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1. C. POWER AMPIFIER

DARKROOM EXPOSURE METEA


FANE 'MODE ONE' HIGH FIDELITY SPEAKER KIT


Incorporating a model $8038^{\prime \prime} 13,000$ Gauss Bass Speaker with ultra low resonance. P.V.C. surround cone. Printed circuit cross-over assembly with ferrite cored coils. Model 303 Pressure Tweeter, Acoustic damping material, Screws, Panels etc., and instructive diagrams, Frequency $\mathbf{4 9 . 9 5}$ post
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$\begin{array}{lll}5 \mathrm{~K} \Omega & 50 \mathrm{~K} \Omega & 500 \mathrm{~K} \Omega \\ 10 \mathrm{~K} \Omega & 100 \mathrm{~K} \Omega & 1 \mathrm{M} \Omega \\ 25 \mathrm{~K} \Omega & 250 \mathrm{~K} \Omega & 2 \mathrm{M} \Omega\end{array}$ $25 k \Omega$
iog or lin less switch ( $81 \mathrm{~K} \Omega \mathrm{lin}$ ) og or lin with sw
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9 volith 20 mA
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| $0 \cdot 1$ |  | 0.25 w | 7p |  |
| 100 | 1 K I2 | 10K! | 100K 2 | 7M S2 |
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| 6000.07 | $0 \cdot 16$ | $0 \cdot 10$ | $0 \cdot 2$ | $0 \cdot 34$ | 0.45 | 1.86 |
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| 1200 - | $0 \cdot 33$ * |  | $0 \cdot 38$ | $0 \cdot 57$ | 0.75 |  |

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U11 25 PNP Sil. Planar Trans. TO-5 like 2N1132, 2N2904
 U13 30 PNP-NPN Sil. Transistors OC200 \& 2 S 104 U14 150 Mixed Silicon and Germanium Diodes
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 U18 86 Amp Silicon Rectiliera BYZ13 Type up to 600 PIV U19 25 Silicon NPN Transistora like BCl08. U20 121.5 Amp Silicon Rectifiers Top Hat up to 1000 PIV. $\begin{array}{ll}\text { U21 } & 30 \mathrm{AF} \text {. Germandum Alloy Transistors 2G300 Beries \& OC71 } \\ \text { U23 } & 30 \mathrm{MADT} \text { \& like MEZ Aeries PNP Transistors .............. }\end{array}$ U24 20 Germansuni 1 Amp Rectiflers GJM Series up to 300 P U25 25300 MHz NPN gilicon Translators 2N708, BSY27. $\frac{\text { U26 }}{30}$ Fast Stilching siticon Diodes like IN914 Micro-Min $\begin{array}{ll}\text { U27 } & 12 \text { NPN Germaniurn AF Transiators TO. } 1 \text { like AC127. } \\ \text { U29 } & 101 \text { Amy SCR's TO E Can, up to } 600 \text { PIV CRSl/25-600 }\end{array}$ U30 15 Plastic Silicon Planar Tranas. NPN 2N2920. U31 20 Siticon Planar Plastic NPN Trans. Low Noise Amp 2N3707 US2 25 Zener Dlodes 400 mW DO-7 case $3-18$ volts mixed V33 15 Plastic Crse 1 Amp Sflicon Rectifers IN 4000 Series U34 90 silicon PNP Alloy Trans. TO-5 BCYa6 2S302/4 U35 25 Sillcon Planar Transistors PNP TO-18 2N2906
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U44 20 Sil. Trans. Plastic TO-5 BC1I5/NPN

## $045 \quad 3 \mathrm{~A}$ sck. TO66 up to 66PI

Code No's. mentioned above are given as a guide to the type of device in the pak. The devices themselves are normally unmarked

QUALITY TESTED SEMICONDUCTORS

|  |  | Price ep |
| :---: | :---: | :---: |
| Q1 | 20 R | - |
| Q2 | 16 White spot R.F. ${ }^{\text {t }}$ |  |
| Q3 |  |  |
| Q4 | 6 Match |  |
| Q5 | $40 C 75$ trans |  |
| Q6 |  |  |
| Q7 | 4 AC 128 tran |  |
| Q8 |  |  |
| Q9 | 7 OC 81 type |  |
| Q10 | 7 OC 71 t |  |
| Q11 | 2 AC 127/128 Complementary pairs PNP/NPN |  |
| Q12 | 3 AF 116 trpe trancistors |  |
| Q13 | 3 AF 117 type transistors |  |
| 14 | 3 OC 171 H.F. type transisto |  |
| Q15 | 7 2N2926 Sil. Epoxy transistors mixed colours . . |  |
| Q16 | 2 anT880 low noise Germanium transigtors |  |
| Q17 | 5 NPN $2 \times 8 T .141$ i $3 \times 8$ T. 1 |  |
| Q18 | 4 MADT $82 \times$ MAT 100 上 $2 \times$ MAT |  |
| Q19 | 3 MADT'S $2 \times$ MAT $101 * 1 \times$ MAT |  |
|  | 121 |  |
| Q20 | 40 O 44 Germanium tran |  |
| 021 | AG 127 NPN Germanium tra |  |
| 22 | 20 NKT trausistors A.F. R.F. coded |  |
| 23 | 10 OA 202 Stlicon diodes |  |
| Q24 | 8 OA 81 liodes |  |
| Q25 | 15 IN914 filicon diodes 75 PIV |  |
| Q26 | 8 OA95 Cermanium diodes sub-min IN69 |  |
| Q27 | 2104 PIV Silicon rectiflers IS 425 |  |
| Q28 | 2 8lifon power rectjuers BYZ 13 |  |
| Q29 | 4 Silicon transletors $2 \times 2 N$ $1 \times 2 \mathrm{~N} 697,1 \times 2 \mathrm{~N} 698 \ldots .$. |  |
| Q30 | 7 Stlicon switch transistors 2 |  |
| Q31 | 6 Silicon sritch transistors 2 N |  |
| Q32 | 3 PNP Silicon transistors $2 \times 2$ <br> $1 \times 2 \mathrm{N1132}$ |  |
| Q33 | Silicon NPN transistors $2 \times 171$ |  |
| Q34 | 7 silicon NPN transistors 2N23 500 MHz (code P397) |  |
| Q35 | 3 Siltcon PNP TO-5: $2 \times 2 \mathrm{~N} 290$ $1 \times 2 \mathrm{~N} 290 \mathrm{n}$ |  |
| Q36 | 7 2N3046 TO-18 plastic 300 MHz N |  |
| Q37 | 3 2N8053 NPN stlicon transistors |  |
| Q38 | 7 NPN transistors $4 \times 2 N 3703,3 \times$ 2N3702 | $0 \cdot 8$ |

## GLECTRONIC SLIDE-RULE

The MK glide Hule, designed to simplify Electronic calculations features the following scales:Conversion of Frequency and Wavelength. Etactance and Self Inductance. Area of Circles. Reactance and Self Inductance. Area of Circles. Volume of Cylinders. Resistance of Conductors Angle Functions. Natural Logs and ' $e^{\prime}$ Functions. Multiplication and Division Squaring, Cubing and Square Roots. Conversion of kw and Hp. A must for every electronic engineer and enthusiast. Size: $2 \mathrm{~cm} \times 4 \mathrm{~cm}$. Complete with case and
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UIC $20=12 \times 7420$ UICE
UIC $30=12 \times 7420$
$=7430$ UIC $40=12 \times 7440$ UIC41=5×7441 UTG42 $=5 \times 7442$ UIC43 $=5 \times 7443$ TIC44 $=5 \times 7444$ $\begin{array}{ll}\text { UTC45 }=5 \times 7445 & 0.50 \\ 0.50\end{array}$ Packs cannot be sylit, but 25 assorted pieces (our mix) is available as PAK UIC X1. BI-PAKS NEW COMPONENT SHOP NOW OPEN WITH A WIDE RANGE OF ELECTRONIC COMPONENTS AND ACCESSORIES AT COMPETITIVE PRICES|8 BALDOCK STREET (AIO), WARE, HERTS. TEL. (STD 0920) 61593.
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| SHURE M31E | 11.63 | 6.95 |
| SHURE M32E | $10 \cdot 73$ | $6 \cdot 50$ |
| SHURE M44-G | $8 \cdot 30$ | $4 \cdot 95$ |
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| SHURE M-44C | 7.90 | $4 \cdot 90$ |
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| SHURE M75-6 (S) | $8 \cdot 36$ | $5 \cdot 45$ |
| SHURE M75-6 | $13 \cdot 60$ | $7 \cdot 60$ |
| SHURE M75EJ | $15 \cdot 40$ | 8.75 |
| SHURE M75ED | 19.00 | 10.15 |
| SHURE M75E/95G | $20 \cdot 80$ | 12.95 |
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| SHURE V15-11 | $39 \cdot 40$ | 24.95 |
| *SONOTONE 9TAHC Dlam/Saph | aph 3.75 | $1 \cdot 45$ |
| Starred cartridges above are ceramic. All others are magnetic. |  |  |
| BASES AND COVERS |  |  |
| CONNOISSEUR Plinth BD1 | 5-57 | 4.50 |
| CONNOISSEUR Cove | $4 \cdot 17$ | $3 \cdot 40$ |
| CONNOISSEUR P.U. Mount Board | ard 0.95 | 0.75 |
| GARRARD WB1 Base | $3 \cdot 78$ | 2.95 |
| GARRARD WB4 Base | $5 \cdot 60$ | 4-25 |
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| Special Offer of Base and Cover to fit GARRARD SP25, SL55, SL65B | $\begin{aligned} & \text { r to } \\ & 65 B \end{aligned}$ |  |
| and 3500 ... | ...Sp. Price | 3-60 |
| GOLDRING PIInth 75 | $9 \cdot 77$ | 6.75 |
| GOLDRING Plinth 72 | $9 \cdot 42$ | $6 \cdot 50$ |
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| C60 | $0 \cdot 71$ | 0.40 |
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| BLANK CARTRIDGES |  |  |
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288-1

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Handling the total of 3000 watt ( 3 kw ) thls unit is unlque for its price in that not only bass, middle and treble but also master controls are provided. Two ampliffer sockets ellminate the need for split leads etc. Supplled in tough white steel case with a blue
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## The Buyer and VAT

During these last few weeks before Value Added Tax takes effect, readers will be contemplating whether to buy now or wait until after April 1st. It will undoubtedly pay to wait on such items as domestic receivers, some $\mathrm{Hi}-\mathrm{Fi}$ equipment and electrical goods, but component suppliers are being hard pressed to meet demand before the infliction of VAT.

One way of circumventing purchase tax has been to supply equipment in kit form, but these items in any form will be liable for VAT at one common percentage rate, since VAT is a tax on the basic purchase price or charge for service.

We would expect that all suppliers will make it quite clear to all customers, at the trading premises and through advertisements, whether quoted prices include VAT. The basic retail price should remain the same as now. Whatever you pay for goods or services from April 1st, you should not be paying any purchase tax or selective employment tax levies; you should be paying VAT on top of the normal retail price, based on the Chancellor's Budget announcement on March 6th.

Kit and component prices should not be raised before April 1st and manufactured goods subject to purchase tax of more than $15 \%$ should be noticeably reduced after April 1st.

Shopkeepers and other traders are not legally bound to show the amount of VAT in cash transactions; neither will they necessarily show special "tax invoices" for retail sales, but it is worth remembering that if you purchase at a cut price rate, the VAT charged will be correspondingly lower. Goods obtained by hire purchase or credit agreements will be subject to tax at the current rate at the time when transaction documents are signed. Interest rates on these terms are not taxable. Private sales between individuals do not attract VAT because neither party is defined as a "taxable person", unless the seller is a trader required to register with H.M. Customs and Excise.

Our advice to readers who are thinking of buying goods by mail order during March is to determine first whether purchase tax is chargeable; if it is then wait until April. Items not subject to purchase tax (such as components and secondhand or used goods) should be ordered early. If of substantial value, send your order by P.O. "Recorded Delivery" and keep the counterfoil of your Postal Order, Money Order, or cheque.

You should be prepared for a VAT levy on charges for packing service but not for postage.
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THE MAY ISSUE WILL BE PUBLISHED ON APRIL 6th

[^3]

## VAT

To the best of our knowledge, the prices quoted in this issue were correct at the time of going to press.

From April 1st 1973, there will be no purchase tax but a large number of goods will carry Value Added Tax. For further details, see our Leader article.

## Tulip Time Rally

The 1973 "Tulip-Time" mobile rally will be held on Sunday May 6th at Surfleet, 4 miles north of Spalding on the A16.

The Spalding and District Amateur Radio Society will also be operating a Special Activity station, GB3STF on $12 \mathrm{th}-13$ th May from The Grammar School, Priory Road, Spalding, in connection with the 1973 Tulip Festival. Operation will be on s.s.b./c.w. on all bands $160-10 \mathrm{~m}$. with a.m. on 2 m . All contacts will be confirmed by special QSL cards.

Of interest to "Going Back" fans-the Wireless Preservation Society will have an exhibition of vintage radio on show to the public. I hope to be able to publish further details later.

## RSGB "Jubilee Year" President



Photograph by P, M. Fletcher.
This picture shows Dr. J. A. Saxton, D.Sc., PhD, CEng, FIEE, FInstP, the new President of the RSGB. On the left of the picture is Swedish Amateur Lt. Col. Per Anders Kinman, SM5ZD, who received the Calcutta Silver Key award. Sir John Eden, the P.M.G. is on the right.

In his address, Dr. Saxton said he felt very proud that he had been chosen as President of the RSGB for its Diamond Jubilee Year. He also said that he was extremely delighted that Sir John had been able to attend and that all there were greatly honoured by his presence.

Dr. Saxton went on to say that the Society would continue to do all in its power to continue to deserve the Ministry's support.

Per Anders Kinman has been made an Honoury Vice President of the RSGB in recognition of his outstanding service with the IARU Region 1 in which he has served for many years.

The Calcutta Key he received was first presented 20 years ago and it has been awarded annually by the RSGB to the person they feel has made, in any one year, the most significant contribution to international friendship and goodwill through the medium of Amateur radio.

## The Buyer and VAT

continued from previous page
If you order from a current catalogue or advertisement after April 1st, you should add the VAT percentage rate to the basic retail price shown. Then add the estimated or stated charge for postage and packing. If a service is chargeable, such as equipment repair or consultation, VAT will be levied at the standard rate. Trading in secondhand and used goods will also be taxable under VAT.
Finally, good news for P.W. readers: newspapers and magazines will be "zero-rated", which means that no VAT is chargeable on these. Transactions of any kind in the U.K. for the direct exportation of goods and services are exempt from VAT.

M. A. COLWELL-Editor.

## Feb P.W. Cover

Many enquiries have been received for information on the book and log book shown on the front cover of the February issue of $P W$. Both can be obtained from the Radio Society of Great Britain, 35 Doughty Street, London, W.C.1. "How to Listen to the World" costs $£ 2$ post paid. The log book shown is intended for transmitting Amateurs and costs 60 p post paid but another one, intended for listeners, is available at 45 p post paid.

## Cordless Iron

This cordless soldering iron comes from the U.S.A. It is completely portable needing no mains power for operation. It has its own power supply which is automatically charged from the stand which itself is connected to the mains. There is a small light near the tip which is very useful for seeing exactly where you are applying heat.

The iron heats up in under 5 seconds and has an "indestructible" iron-plated bit. It costs £9.25 complete with charging stand and fine bit. A heavy-duty bit is also available.

The exclusive distributors in Great Britain are Pact International Electronics Ltd., P.O. Box 19, Royston, Herts, SG8 5HH.


## Sonex 1973

The technical author and writer Donald Aldous will be heading a small team to organise the Special Features at "Sonex '73".

Two recital/lecture sessions will be presented each day at the Exhibition's new venue the Excelsior Hotel, London Airport, Heathrow.

Speakers will include Bert Webb, who will lecture on pickups, arms and turntables and Kenneth Shearer, speaking on listening room açoustics.

There will also be a lecture on tape cassettes and an examination of current loudspeakers and trends.
Sonex ' 73 will run from Friday, March 30th to Sunday, April 1st, inclusive.

## Sound '73

"Sound ' 73 " is the exhibition run by the Association of Public Address Engineers. New P.A. equipment and new techniques in the public address and allied fields will be on show.
As in past years, there will be a lecture programme held in the hotel's "City Room". John Maunder from Shure Electronics will give a lecture and demonstration on microphones on Tuesday, 13 th March at 14.30 hrs .
Mike Beville from Audio and Design Recordings Ltd. will lecture on Components and Limiters (Wednesday at 14.30 hrs .)
On Thursday March 15th at 10.30 hrs , a member of the Hornsey College of Art will talk on Industrial design of Public Address Equipment.
Venue will be the Bloomsbury Centre Hotel, March 13-15, from 10.00-18.00hrs.

## Jermyn Inverters

Jermyn Distribution market two inverters, both giving an output of 250 V . Input for the 150 W version is 12 V and for the 300 W type, 24 V .

A feature of the design enables these inverters to charge the $12 / 24 \mathrm{~V}$ batteries up to 10A when plugged into a power socket.
If the mains supply fails, the inverter automatically switches to its "invert mode" providing 240 V emergency power immediately.
Should the unit be accidentally overloaded its drive is so adjusted, that the output voltage falls completely, switching the


## I.E.E. Conferences

The following new conferences to be held in 1973 and 1974 are being organised by the Institution of Electrical Engineers.
"Electrical Signals from the Brain" 27-30 August 1973-in Oxford.
"Symposium on Electromag. netic Wave Theory" 8-12 July 1974-in London.
Further details of the conferences mentioned above are available from the Conference Department, IEE, Savoy Place, London. WC2R 0BL.

## De-soldering Unit

In the February issue we mentioned the Litesold De-Soldering Unit and a misprint stated the price at $£ \cdot 75$. This should have read $£ 2 \cdot 75$. Apologies to Light Soldering Developments and any readers that may have been inconvenienced.
unit off, eliminating any problems.
An internal 15A fuse blows should the battery leads be connected wrongly. Indicator lights illuminate to show whether the unit is charging a battery or providing a 250 V 50 Hz output.

Both types of inverter can be obtained in kit form or already built. The kit for 150 W costs $£ 25$ (£29 built) and 300 W £34 ( $£ 39$ built). Jermyn Distribution, Vestry Estate, Sevenoaks, Kent.


VHF radio provides a most useful signal source for any hi-fi enthusiast and; despite the 15 kHz frequency limits, the live BBC broadcasts are certainly up to the standands of the finest records.

The advantages are numerous-being a nonmechanical system the distortion is very low and the self-generated noise is also very low, the result being a large dynamic range and the possibility of encoding two channels on a single signal. This, plus the excellent BBC technique, makes f.m. radio an invaluable source of material. Unfortunately there are several difficulties in securing top quality f.m., regardless of the equipment used.

## SIGNAL-TO-NOISE

Referring to Fig. 1, it can be seen that the performance of a receiver depends on the strength of the signal applied to the aerial socket of the receiver. The mono signal-to-noise ratio assumes an optimum value (usually 50 or 60 dB ) with only a few $\mu \mathrm{V}$ input-even on cheap tuners. The signal-tonoise performance does not improve so rapidly on stereo, however, and in many cases could be improved on even with inputs in the region of a few hundred "V-hence the increase in noise on stereo


First of the two VHF amplifiers described in the articie. This one uses field-effect transistors.
in most situations. A person who gets no increase in noise is indeed most fortunate and need read no further!

As regards crosstalk and distortion, these parameters usually reach optimum values with fairly small inputs and usually there is not much improvement to be had by an increase in the signal.


Fig. 1: Graphs comparing receiver performance with the input signal level.
Very few people are fortunate enough to be able to receive "near perfect" stereo without a large aerial, loft or roof mounted. A large aerial, besides having the necessary gain usually has the advantage that it is highly directional and eliminates multipath distortion from reflected signals. In many situations, a reasonably large aerial, three or more elements, carefully sited, and with a low loss feeder gives good stereo but perhaps not of the very highest quality.
The major downfall in domestic stereo reception is that of "stereo" noise-perhaps not excessive but sufficient to mar an otherwise excellent signal. The obvious way to reduce this noise is to increase the signal feeding the tuner by an "aerial amplifier". At present in this discussion we must assume that the amplifier be perfect, having gain but generating no noise itself.

## OTHER NOISE

At this point many readers may well be wondering about the other forms of noise, thermostats, car ignition and other spurious signals. These signals
are picked up by the aerial and sometimes, even, by the feeder, and consequently will be amplified, together with the f.m. signal, by our "ideal" amplifier, which we are using to combat stereo noise. This brings us to the last important parameter of f.m. receivers, amplitude modulation limiting.

The "ideal" f.m. tuner is designed to respond to f.m. only and a.m. signals are rejected to an infinite degree. Unfortunately, no receiver is perfect and some degree of sensitivity to a.m. is unavoidable.

The a.m. limiting of the receiver improves with increasing signal input, as seen from Fig. 1, so that increasing the signal and spurious a.m. together actually reduces the level of audible interference. Unfortunately, when the receiver is switched to receive a stereo transmission, the stereo decoder in the receiver is actually only sensitive to amplitude modulation of the stereo subcarrier. In the decoder some measure of a.g.c. is applied to the subcarrier amplifier and this maintains a constant level of stereo interference from an a.m. signal regardless of the actual input voltage to the tuner.

It can therefore be seen that whilst increasing the tuner input with an "ideal" amplifier, the "stereo" interference level remains substantially constant whilst the "mono" component of the audible interference has been reduced by virtue of the amplitude limiting of the f.m. discriminator.

## AERIALS

Earlier in this discussion, we talked of elaborate and efficient aerial systems but the previous discussion on aerial signal amplification implies that, with a good amplifier, an aerial providing only a minimal "clean" signal can be used to run an f.m. tuner with a high degree of success.

Unfortunately, some tuners are not so sensitive as others and sometimes require millivolts to function at their best. This means considerable amplification of the few tens of microvolts coming from a rather poor aerial!

The better tuners nowadays use f.e.t.'s in the input circuits to keep noise to a minimum and to keep cross modulation low; so any proposed amplifier should have at least as good a performance as the front end of the tuner.

This is very demanding when tuners such as Sony, Radford and Fisher are considered, but it is possible to improve these tuners on a very poor signal by a remarkable degree.

. Fig. 2: Circuit of the amplifier using two f.e.t.'s.

To design an amplifier with a front-end comparable to the best tuners use has to made of f.e.t. devices which possess the best noise properties of all the common solid state devices presently available and are responsible for the excellent performance of the best tuners.


Fig. 3: Details of simple transformer to match balanced $400 \Omega$ feeder to $75 \Omega$ coaxial feeder.
The design chosen for the amplifiers in this article is not often used with solid state devices but was once very common with valve Band III television tuners. The amplifier is a cascode configuration giving reasonable gain with very low noise and good cross modulation.

The input to the amplifier, Fig. 2, is 750 co-axial feeder so Fig. 3 shows a transformer that can be constructed from a small ferrite bead for $400 \Omega$ balanced feeder.

## THE PCB

The amplifier can be very easily built on a single piece of printed circuit $4^{\prime \prime} \times 2^{\prime \prime}$ and it is recommended that fibreglass board be used.

To etch the required print design as shown in Fig. 4 (full size), the diagram is traced on to the copper side of the print with carbon paper and the marked areas painted. in with high quality enamel. When the paint is dry, the print is immersed in the etching solution and agitated until completely etched (only paint and base material can be seen). The etching solution can be ferric chloride solution but a useful alternative is "Liquor-Ferri Perchloridi Fort" which is obtainable from chemists and shouid be used undiluted. Rub the copper surface of the board with "Ajax" or similar scourer before painting, to ensure a clean surface.

The enamel paint can be removed with paint stripper once the etching is complete and the board washed and cleaned with wire-wool. The holes for the components should be drilled with a $t_{32} /{ }^{\prime \prime}$ drill, preferably in a wheel brace.

Construction is straightforward except that the f.e.t.'s should be left until last and soldered in with the iron disconnected from the mains-surprisingly little voltage will break down the gate junction.

The coils, Fig. 5, are wound on the formers which are located on the board and stuck with Araldite. The 30 pF variable capacitors must be connected so that the rotor is connected to the earth line. The author used 30 pF Philips beehive trimmers mounted



Fig. 5: Winding information for 11 and 12.

## components list



Fig. 4 : Printed circuit board, for the circuit of Fig. 2, shown actual size.
on the copper side of the board. The resistors should be small low noise types and the capacitors need only be low working voltage.

The completed amplifier should be screened by placing tin plate at the position indicated and soldered to the positive line as support.

The amplifier is mounted in a small wooden box, measuring approx. $4^{\prime \prime}$ cube internally, covered inside with tin foil, which is connected to the positive supply line. A front panel can be made from a piece of double-sided copper board with the input and output sockets and the on/off switch. The printed board is soldered to the front panel by the positive line, so earthing the front panel.
With a 12 V battery connected, the output is fed to the tuner and the input is fed by an "inefficient" aerial, in most cases a short piece of wire.

With the amplifier on, the tuner is switched to the local stereo station and the tuning capacitors C3 and C7 set midway. The cores in the coils (Ll and L2) are then adjusted to give the strongest signal, best observed on a tuning meter or audibly by judging least noise.

The amplifier is then assembled in the case and the capacitors C3 and C7 tuned to give the best signal. The tuning will be sharp and critical, so care should be taken when adjusting.

The amplifier, when connected between the proper aerial and tuner, should have sufficient gain to overcome most of the tuner noise, without introducing significant noise itself. When used on weak local radio stations, the improvement will be very marked, especially in mono. Unfortunately, this usually requires the amplifier to be re-tuned to the local channels. However, connecting $27 \Omega$ resistors across L1 and L2 will broaden the bandwidth considerably, enabling the amplifier to be used over a number of channels.

## BF200 AMPLIFIER

However, for many of the "poorer" tuners, this amplifier will still not be good enough to fully drive the tuner on a moderate signal. For optimum stereo in such cases two of the amplifiers could be


Fig. 6: Circuit of the second amplifier, using BF200 transistors.


Fig. 8: Method of mounting feed-through


Fig. 7: Illustration, actual size, of the printed circuit board for the BF200 amplifier.
connected in series. The price of f.e.t.'s, however, makes this uneconomical and ordinary low-noise silicon transistors used after the f.e.t.'s work very well. The proposed amplifier, Fig. 6, is two stages of a conventional r.f. amplifier design. Fig. 7 shows a full size print layout for two stages, although for additional gain or bandwidth more stages can be used, although greater than five would be difficult to keep stable.

The board is ctched in the same manner as before and a similar front panel and box are prepared. Alignment is the same, tuning the coils and finishing with the capacitors.

The transistors specified are BF200 and these should be used for at least the first two stages to keep the noise level down; after this such transistors as 2 N 706 work very well. An alternative for the BF200 is $\operatorname{BF} 180$ or BF181.

Fig. 8 shows how C5 and C9 are constructed with 1000 pF feed-through capacitors. L4 is made by wrapping a few turns of $36 \mathrm{~s} . w . g$. d.c.c. copper wire round a ferrite bead. Fig. 9 gives the winding information for L1, L,2 and 1,3.

The two amplifiers just described will enable good stereo to be obtained from a mediocre signal without the need for a "fancy" tuner. The improvement that can be expected depends greatly on the individual circumstanccs. The author has used the amplifiers with a Sony ST5100 tuner and a poor aerial and found the results almost as good as when a proper aerial was correctly installed on the roof. The cost of a pair of amplifiers is comparable to a decent


Note:
L. 1 L. 2 \& L. 3 wound on $\frac{1 "}{4}$ Dia formers. L. 1 use 22 SWG copper wire. L2 \& L3 use 18 SWG copper wire.

TB 63

Fig. 9 : (above) Detalls of the three coils shown in Fig. 6. In the photograph (below) the second amplifier board is shown soldered to the front panel consisting of double-sided p.c.b.

aerial less downlead and fitting, so financially it is a viable proposition, although not a substitute for a first class aerial.

In order to split one signal to run several outlets and obtain maximum signal at each of the outlets, a star-splitter is used. This arrangement provides correct impedance matching throughout the system.

Figure 10 shows the circuit of a star-splitter. The value of the resistors $R$ depends upon the impedance of the feeders ( $Z$ ) usually $75 \Omega$, and on the number of outlets ( $n$ ) which can be as many as required.


Fig. 10: Circuit of star-splitter enabling several outputs to be obtained from a single input.

The relationship is:- $\mathrm{R}=\mathrm{Z}(\mathrm{n}-1) / \mathrm{n}+1$. For example, for a splitter for a $75 \Omega$ system with two outputs the resistors to the nearest preferred value are $27 \Omega$. Small ${ }_{4}{ }_{4} \mathrm{~W}$ resistors are suitable and indeed preferable to physically larger types.

All of these systems involve a loss of signal and this must be made up by amplification before the splitting. For a distribution system, splitting six or


Fig. 11: Suggested power supply for the v.h.f. amplifiers.
eight ways, two f.e.t.'s followed by three transistors should be adequate. One important point to note is that the outputs of the splitter must all be loaded by a $75 \Omega$ resistor when not in use, otherwise a loss of signal results and the splitter may form the node of standing waves in the feeder.

If either of the amplifiers is built as a distribution system, it is perhaps a good idea to build a small mains unit to power the whole system. It is desirable that this supply be stabilised to guard against changes in the tuning of the f.e.t.'s if the supply voltage changes.

Fig. 11 gives a circuit of a simple power supply based on an 8 V bell transformer with voltage doubling and zener stabilisation. The lamp, LP1, can be used to give on/off indication.

It is hoped that this article has given sufficient information for the home constructor to build an amplifier system to improve the performance of a tuner to reach the very high standards of the B.B.C. stereo transmisions.

## EXPEAIMENTERS CORNER

## NOVEL TONE GENERATOR

WHENEVER home constructors and experimenters need to build a source of audio frequency, the choice nowadays is usually a transistor multivibrator. The reasons for this selection are easily understood. It requires few components, avoids the need for a transformer and gives a strong output signal. These virtues outweigh the fact that the square-wave output does not make for a particularly pleasing sound when amplified and fed to a loudspeaker.

In designing the a.f. tone generator to be described, the author set out to devise a circuit which would possess all the aforementioned merits of the multivibrator and at the same time give a better audio tone. In the circuit shown in Fig. 1-which, as a transistor circuit, is thought to be original-the writer believes these objectives have been largely achieved.

## N. NAUGHTON

It will be noticed that the circuit is actually simpler than a multivibrator and basically it is a sawtooth generator. Like all sawtooth generators it depends upon the charge and discharge times of a


Fig. 1 ; Circuit of tone generator.
capacitor for its action. But as linearity of the sawtooth ramp is not important for the required purpose, the provision of constant current is omitted in the circuit shown.

In Fig. 1 the two transistors perform the function of switches and the circuit operates as follows:capacitor $C$ charges via the charging resistor Rc and the base-emitter junction of Trl. The charging current flowing through the $B-E$ junction turns on Tr 1 and the flow of collector current causes a voltage drop across Rl which results in the potential on the base of $\operatorname{Tr} 2$ dropping to a level low enough to turn off $\operatorname{Tr} 2$. As the charge on C rises the charging current diminishes so that the potential on the base of $\operatorname{Trl}$ decreases to a point where $\operatorname{Trl}$ turns off. The base of Tr2 now goes strongly negative causing Tr 2 to turn on and C now discharges via Tr 2 into the discharge resistor Rd. While discharge current is flowing in Rd it maintains a positive potential on the base of $\operatorname{Tr} 1$ thus preventing further switching action during the period of discharge. When $C$ has discharged, it again draws current through Tri once again turning off $\operatorname{Tr} 2$. The cycle then repeats.
The foregoing is a simplified description of the operation but many will find it interesting to measure the balance of currents and potentials which exist in the circuit under static conditions when $C$ is removed.

The frequency of oscillation is not readily calculated since it is influenced by several factors one of which will be determined by the type of coupling employed to connect the output. However, for practical purposes values of C ranging between $2 \mu \mathrm{~F}$ and 5000 pF will cover the audio spectrum.

## APPLICATIONS

A useful feature of the circuit is that it uses only one capacitor. With the values given, the generator should oscillate at any point within the a.f. range the actual point being determined by the value of $C$ selected. Thus it becomes a simple matter to construct a signal generator giving a number of spot frequencies as desired; the only extra components needed being a multi-pole switch and a few capacitors. Many other audio applications will doubtless occur to readers.

For those interested, an interesting feature of the circuit is that because the charge and discharge paths of $C$ are independent, the circuit can readily be made to produce right-handed or left-handed sawtooth waveforms. Moreover, by choice of p.n.p. or n.p.n. transistors, the polarity of either slope can be created as desired at source, thereby eliminating inverting stages. Although some value of Rd is essential for circuit function, in practice this can be less than $100 \Omega$ thus allowing fast flyback times. There is no reason why $C$ can not be a very large value electrolytic so that the circuit can easily be adapted for timing application.

## GENERAL

No constructional details are given since the generator will mostly be incorporated in some larger project. Experimenters can assemble the components on a length of Veroboard or tagboard.

As stated at the outset, the circuit is primarily intended as an a.f. generator and the values given have been chosen to give a roughly pyramid waveform at the collector of $\operatorname{Tr} 2$. A strong output signal is also available at the collector of $\operatorname{Tr} 1$ but this is a
steep sided pulse and the "softer" toned signal at the $\operatorname{Tr} 2$ output can be readily appreciated. Coupling of the output is perhaps best determined to suit the individual case but too tight a coupling to the Tr2 output is best avoided. This is less important if the Trl output is used.

With the component values given, the circuit places no critical demands on the transistors' characteristics. Seven widely differing types, in various combinations, have been tried with success. However, experimenters using low values of Rdespecially with a high value of C-should note that $\operatorname{Tr} 2$ must then be a device capable of handling the heavier discharge current. Tr 1 and Tr 2 must both be either p.n.p.'s or n.p.n.'s, the -9 V becoming +9 V if n.p.n.'s are used.

## TETEUSTIII <br> APRIL ISSUE <br> TOUCH TUNING <br> The latest development in television receiver design is fouch-sensitive tuning, using touchbutton units which provide a completely nonmechanical means of chammel changing. A very high-impedance electronic switching circuit selects the required channel when a finger is placed across apart of contacts to complete the appropriate, circuit. Several models featuring touch tuing are now on the market and this month, we are investigating the technique and the circuitry involved, The change from mechanical to all-electronic channel selection should improve the stability and rellability of TV tuning.

## BAND UHIRREAMPLIFIER

Another Bunney, wideband aerial amplifiec Intended for $D X$ Work The wo transistor (BF272) design on testifted, weak signal from mere traces of lie sync pulses 10 a solldly locked. nolsy mage Patiof Rógers pesentprogramm


SERVICING
elentyon thes servicing font this month gobn Laven his new faul, finding series investigates the lie timebase used in mosi KB-STC models. Vivian Capel completes his guide to power supply ctrcuit servicing. Les Lawry-Johns starts on the single-standard Bush-Murphy TV181SI V2016 series. And. Caleb Bradley takes us further with the BRC 2000 colour chassis.

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## D.SILVESTER

THE audio amplifier to be described was designed to deliver in excess of 10 W r.m.s. into an $8 \Omega$ load, whilst being constructed in a $41_{2}$ in. $x$ $3^{1} 2 \mathrm{in} . x$ 2in. diecast aluminium box. The size of the box is, in fact, limited not by the size of the i.c. but by the passive components, especially the output capacitor which must be included in the circuit. The use of the i.c. has added one inestimable advantage in that the current through (and the power dissipated by) the output stage under overload conditions can be limited. It is thus possible for the output to be short circuited without disastrous results.

The i.c., although capable of giving a 1 watt output without any other transistors, has been primarily designed for use as the driver stage of an amplifier
capable of delivering up to 35 watts into an $8 \Omega$ load.
A description of the working of the i.c. has not been included as it is not necessary for the constructor to completely understand the operation of the device for it to be used successfully.

Two transistors in the i.c. are responsible for the current and power limiting facility. It was calculated that under maximum current limiting and with the normal current gain of a power transistor, the output stage of the i.c. would be overloaded. One way of curing this fault is to use a Darlington transistor. This in fact was done, the Darlington pair being contained in the TO3 metal case of the power transistor. The current gain of the complementary Darlington pairs is in excess of 1000 .

## CALCULATIONS

The final circuit used for the prototype amplifier is shown in Fig. 1 but since other constructors may wish to use different power supply voltages and load impedances from those used by the author, the method of calculation of the current and power limiting components is given. The basic circuit should be adhered to as it is only the values of R6, R8, R9 and R11 which need to be changed.

The following section shows the worked equations used to give the value of the components in Fig. 1.


Fig. 1: Circuit diagram of the amplifier and pin connections for the output transistors.


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Fig. 3: Component layout on to $\mu$ of veroboard, the copper strips running horizontally underneath.


Fig. 2 : Drilling details for the box lid on which components are mounted
They are based on a supply voltage ( Vcc ) of 35 V and a loudspeaker impedance of $8 \Omega\left(\mathrm{R}_{\mathrm{L}}\right)$. Before beginning the calculations it must be remembered that there are two limitations imposed by the active devices used; the supply voltage must be in the range 10 to 40 V , and the maximum current through the transistors $\operatorname{Tr} 1, \operatorname{Tr} 2$ must not exceed 4 A . We can calculate:


This view of the completed amplifier emphasises the restrictions on the physical size of the output capacitor C6.

Peak current in transistors is:

$$
\frac{\mathrm{V}_{\mathrm{cc}}}{2 \mathrm{R}_{\mathrm{L}}}=2 \cdot 2 \mathrm{~A}\left(\mathrm{l}_{\mathrm{pk}}\right)
$$

If this value is above 4 A we must use either a lower supply voltage or higher load impedance. 'To provide current limiting of this value and assuming no voltage drop across R7, R10 we calculate the values of R6 and R11 as:

$$
\mathrm{R} 6=\mathrm{R} 11=\frac{650 \mathrm{mV}}{\mathrm{I}_{\mathrm{pk}}}=\frac{0.65}{2.2}=0.296 \Omega
$$

The nearest preferred value of resistor, which must be lower in value than the calculated value, is $0 \cdot 27 \Omega$. The basis of the working of the power limiting facility is difficult to explain in such a short article but works on the principle that both a high voltage across or a high current through a transistor and its network of resistors (Tr1, R6, R7 and R8) will tend to turn off the transistor. The resistor network acts to sense the power dissipation in the transistor and to limit this dissipation. The value of R 7 and R 10 is $56 \Omega$. In all cases the value of R8 and R9 is:

$$
\mathrm{R} 8=\mathrm{R} 9=\frac{\mathrm{V}_{\mathrm{cc}}}{6 \mathrm{~mA}}=5.8 \mathrm{k} \Omega
$$

The next highest preferred value, $6 \cdot 2 \mathrm{k} \Omega$, is rather difficult to obtain and the value has been increased to $6 \cdot 8 \mathrm{k} \Omega$. The voltage gain of the circuit is 100 and although this may be altered by changing the value of R4, the effect of such a change on the stability of the amplifier is difficult to predict. Under d.c. conditions the capacitor C 3 represents an open circuit and the d.c. gain becomes zero, the voltage at the emitters of the two transistors, determined by the ratio of the values of R1 and R2, being at half the supply voltage.

## components list

## Resistors

| R1 4.7 kS | -R5 Joks | R9 $6 \times 8 \mathrm{kN}$ |
| :---: | :---: | :---: |
| R2 $4.7 \mathrm{k} \Omega$ | R60 27 + 02 | R10. 560 |
| R3:10kS | R7 560 | R11.0.270-02 |
| R4 1002 | R86880 | VR1 5kn log |
|  | A1 $\frac{1}{2}$ W $5 \%$ excep | R6and R11 |

## Capacitors

C1 $1-64 F 40 \mathrm{~V}$
C2 : 6. 4uF 25 V
c4. 10 pf
C5 2000pF
C6 500 mF 50 V
C3 $50 \mu$ F 35 V

Note: C 6 should net be over $2 \frac{1}{2}$ It Iong and 1 , diat.

## Semiconductors

Tri Md4000: $\because$ TH2 M14010
(both with mica washers and bushes)
1CI NE540L
Available from SCS Components, POB 26 , Wembley, Middesex

Miscellaneous
Diecast box $4 \frac{1}{2}^{\prime \prime} \times 3 \frac{1}{2}^{\prime \prime} \times 2^{\prime \prime}$. Veroboard $29^{\prime \prime} \times 2 \cdot 2^{\prime \prime}$, $0.1^{\prime \prime}$ matrix. mput socket Outpuf jack socket.

## CONSTRUCTION

The complete circuit is constructed in a diecast aluminium box using the lid'to hold all the components and the rest of the box as the cover. Fig. 2 shows the lid marked out for drilling, after which the power transistors are mounted (the NPN transistor Tri, MJ4000 on the left as seen in .Pig. 2) using mica washers with bushes. It is best at this stage to check whether the collectors of the transistors have been accidentally shorted to the box lid. The remainder of components should be fixed to the box lid, a tag washer under the input socket nut providing the main connection to the lid.

The component board is a $2 \cdot 9 \mathrm{in} . \times 2 \cdot 2 \mathrm{in}$. piece of $0 \cdot 1$ matrix veroboard, with the copper strips running along the $2 \cdot 9$ in. length of the board, Fig. 3 . After cutting the strips below the i.c., drilling the mounting holes and cleaning up the rough edges, a few moments should be spent with a meter checking that no lines are shorted together and that the strips below the i.c. are completely severed.
There should be no difficulty in constructing the board, but it is advisable to use different coloured wires for the flexible connections as this assists in the final wiring. The capacitor C6 is held in place by its wire leads, the positive end of the component being connected to the emitter of Tr1. The circuit board is held above the transistors on four 4BA screws with either two sets of nuts and washers or lin. long spacers. The nut heads are stuck on the box lid with epoxy resin adhesive.
The final stage of wiring consists of connecting all the leads from the board to their respective terminations, wiring a negative supply earth lead to the earth end of the output socket, the negative supply socket, and the solder tag on the input socket. Lastly a wire is taken from the centre of the input socket to one side of VR1, the remaining potentiometer connection being taken to earth (the slider goes to the input on the board).
A power supply capable of delivering the maximum requirements of the amplifier must be used. In the author's case a supply of 35 V at $2 \cdot 5 \mathrm{~A}$. Under these conditions the amplifier was able to deliver nearly 15 W r.m.s. into an $8 \Omega$ load.

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## ON RECENT DEVELOPMENTS

FOR some time now, weather satellites have been happily orbitting the earth some 800 miles up. These animals transmit a picture of the earth and a suitably equipped station below can receive these pictures. They show cloud distribution which is useful in helping to guess whether it will rain or not on the following washday.

## 3D RAIN?

Now comes news about a satellite built by the RCA. It is equipped with special sensors and boasts the distinction of being the very first operational satellite to provide a three-dimensional picture of the earth's weather. It is equipped with some very impressive sounding equipment; a scanning radiometer, a very high resolution radiometer and a vertical temperature profile radiometer.

The very high resolution radiometer (v.h.r.r.) has two channels and offers images from both the visible and infra-red regions. It can sense the thermal energy given off by objects below, down to half a mile in area.

Resolution is far better than the pictures provided by television cameras or the radiometers used in earlier satellites. The solar cells in the satellite, NOAA-2, generate some 500 watts and over 10,000 individual cells are used to achieve this power. Who knows-television announcers of the future may be forecasting that three dimensional rain is on the way.

## LIQUID CRYSTALS

It now seems a certainty that liquid crystal displays will have a big impact on the electronics market. 1 reported these as being already built into a multi-digit display being marketed at the Electronica exhibition in Germany late last year. Now, the U.S. giant, North American Rockwell is reported to be mass-producing liquid crystal devices.

The liquid crystal phenomenon was first discovered way back, as long ago as 1888. The basic principle is that certain liquids become opaque when they are subjected to an electric field. With no field present, they are transparent. The North American Rockwell displays consist of two small plates of glass. Although
these are bonded together, they are separated by a tiny gap-about a thousandth of an inch. This tiny gap is filled with the special liquid which is affected by an electric field. The cell is thus completely transparent.

However, when a small electric field is applied (this takes about 25 V to achieve the desired effect) the cell becomes opaque. Without a field, the molecules in the liquid are uniformly and regularly oriented. The application of the electric field causes this molecular structure to become disoriented and the molecules thus scatter the light giving an opaque effect.
The voids or spaces into which the liquid is put can be formed into figures or letters for the purposes of forming a display. Conductors to the liquid are formed from tin oxide which is a transparent conductor. The liquid crystal display draws almost no power-about 10 mW or one hundredth of a watt. Perhaps we will have windows some day which will be clear until we press a button?

## OZONE SPEEDS UP RESIST REMOVAL

Integrated circuits (i.c.'s) are delicate little beasts especially during their manufacture. Part of this process is a photographic one. There is a particular stage in the printing of each mask where unwanted photo-resist has to be removed. Considering the size of elements on individual i.c.s this removal is a delicate business. The General Electric Company in the U.S.A. have come up with a new method of removing unwanted photoresist. It consists of exposing the resist to ultra-violet rays in the presence of air. It takes about 40 minutes for a typical film of resist to be removed. The depolymerisation of the resist happens at a rate of about a thousand angstroms per minute. By injecting 2\% ozone gas into the chamber with the specimen the process is speeded up some ten times.

Earlier, I talked of liquid crystals and many have leapt in claiming that the solid state flat tube television screen is here. One British manufacturer, however, is obviously unimpressed. Mullard has opened a
new factory in County Durham. The sole purpose is to manufacture shadow mask colour television tubes.

Although only officially opened in December last year, pilot production was already said to be running at the rate of 100,000 tubes delivered from this plant. By the end of this year, production is estimated to be in the region of 500,000 tubes and by the end of this present decade, production will be running at some 900,000 tubes/annum. All this adds up to a lot of colour television receivers-have you got yours yet?

## LASERS ARE BACK

Lasers are back in the news again, this time in America where the Optical Physics Division of the National Bureau of Standards is applying one to monitor pollution. A major contributor to atmospheric pollution is nitrous oxide. It is found lurking in such jolly jollop as flue gases and in the exhaust fumes from motor vehicles.
The laser beam is directed through a magnet whose coils are driven by an audio amplifier and is detected by an infra-red device. The amplifier generates a modulating magnetic field which has the effect of sweeping the absorption line through the laser line. The detector produces a signal which has a direct relationship to the amount of nitrous oxide in the sample being investigated.

Concentrations of nitrous oxide as low as 20 parts per million (p.p.m.) can be detected but workers believe that concentrations right down to 0.1 p.p.m. will be detectable as the method is improved.

## FLYING SPOT

Useful things oscilloscopes. Have you ever wondered just how fast that tiny spot of light travels across the cathode ray tube? Before you start to make calculations, think of the latest fast writing oscilloscope, the HP 184A. The writing speed for this unit is so fast that the spot travels 400 cm in one millionth of a second.
Cimbers


EVER since the day when Marconi equipped the yacht Elettra with radio to act as a mobile test-bed, electronics and boats have been closely associated. The first really extensive use of radio was at sea, and after the Titanic disaster all ships were required to keep radio watch continuously for distress signals. The requirement for ships to be fitted with radio and staffed with trained operators did not, however, extend to small boats at that time, and it is only fairly recently that radio and electronic aids for pleasure craft have become more of a necessity for serious boating.

## MARINE HAZARDS

Electronic devices for use at sea present some design problems which do not appear in land-based equipment. The most obvious hazard is moisture, not the fresh water moisture of a foggy day on land but the salt-laden moisture which corrodes its way through soldered joints, cable terminations, plugs and sockets, and most other metal-to-metal joints in a remarkably short time. As well as the corrosion problem, which can be completely solved only by total encapsulation, there is the problem of insulation resistance.

Salt spray is quite a good conductor and in valve equipment in particular, operating at high voltages, protection of equipment from saline elements is a problem to designer and operator alike.

It was largely the water-spray problem, as well as the cost factor, which has kept electronics out of smail craft for so long. In large vessels, the radio and radar equipment could at least be mounted in an enclosed space at a reasonable height above the deck, well away from the direct spray except in the most extreme conditions. In small boats, such sprayfree havens cannot be found, and there is little chance of keeping the equipment free of water by heating coils, as were once used on bridge-mounted radio and radar.

The advent of solid state equipment has opened up a new dimension in marine applications, especially with the possibility of cutting down on interconnections by using integrated circuits. Vibration is one of the most pressing problems in small craft, where the rumble of even a small diesel engine is enough to loosen plugs from sockets or wires from soldered joints.

Some of the most up-to-date equipment used for pleasure craft can be seen at the annual Boat Show, held each year in London in January. This year you *would have found amongst other more glanorous displays, several kinds of communications equipment.

One stand showed a "triple-standard" television receiver, complete with inverter for use with ship batteries, for use in most yachting centres in the world. But for those who insist on wind in their sails, there was a complete analogue computer, taking information on wind speed and direction, water speed, currents, intended course, and displaying the optimum trim on a model.

## RADIO AIDS

Radio is still the mainstay of marine electronics, and it is hardly surprising that there should not be many advances in this direction as there are in instrumentation. The latest ranges of transmitter/ receivers are geared to the most recent International regulations, which call for s.s.b. working in the marine bands between 250 kHz and $4 \cdot 1 \mathrm{MHz}$ and the use of all solid state equipment nas resulted in very compact apparatus.

A large number of v.h.f. and u.h.f. radiotelephones were also shown, and one interesting new development is the Marconi "Pocketphone" intended for communication on large vessels, such as tankers, and during docking and cargo handling operations.

The v.h.f. and u.h.f. equipment uses frequency modulation, and most make provision for the reduction of channel spacings to 25 kHz , which will soon be a Post Office requirement. The bands used are within the range 148 MHz to 174 MHz in v.h.f. and 440 MHz to 470 MHz in u.h.f. Communications equipment was being shown by (among others) Marconi Marine, Electronic Laboratories (Marine), Ajax Electronics and S.P. Radio A/S of Denmark.

The latter firm was showing a receiver operating under the spray from a waterfall, whose salt content I did not sample, and also showed the "Dual-watch" facility on receivers. This enables the operator to fulfil the requirements of listening for distress signals along with normal working. The distress frequency is preset, so that the set is tuned permanently to that frequency in the watching circuits.

Working can be continued on other channels, and if a signal is received on the distress frequency (or any other being watched) the receiver output is instantly and automatically switched to the watched frequency. A lamp indicator also shows when a call is being received on the watched channel, and flashes during normal operation to show that the watch is being carried out.

## DIRECTION INDICATOR

An interesting device for the small boat is the "Seafix", a hand-held radio direction indicator by Electronic Laboratories (Marine). This consisted of a t.r.f. transistor receiver with "autodyne" detection
over the range 200 to 400 kHz , with the calibration most accurate at 300 kHz , the centre of the Marine Beacon band.

Switched to NAV, the autodyne circuit is used to receive beacon transmissions, and the highly directional ferrite rod aerial can be used for finding the bearing of the beacon within $2^{\circ}$ (at half rated range). Switched to BC, the "Seafix" can be used to pick up the long-wave shipping forecasts broadcast on this band. A stethoscope headset was used in preference to loudspeakers or conventional phones, and water sealing was by the conventional compressed rubber gaskets.

## RADAR

Radar has for a long time been one of the most important navigational aids for larger vessels, but is seldom seen on smaller craft. There are several reasons for this: the cost of any radar installation is high, since it must include the cost of modifications to the vessel as well as the provision of the apparatus.

Radar equipment is often unavoidably large, heavy and dependent on high voltages, since the cathoderay tube and the magnetron must use high voltages, although it is now possible to replace the klystron local oscillator by Gunn diodes or other semiconductor devices.

The Hepplewhite Marine Check radar detector lets you know if you are on a collision course.


Manufacturers are not dragging their heels over these problems, but are commendably cautious, since they know only too well how much the readings of a radar receiver come to be relied on. Any compression of size or cost must therefore not be made at the expense of accuracy of plotting if we are not to have a rash of "radar-assisted" collisions. All the big names in marine radar (Decca, E.M.I., Marconi Marine) showed small radar outfits, but these were not really intended for the amateur.

The Hepplewhite "Marine Check" is a continuous scanning self-powered instrument designed to minimise risk of collision with other vessels whose radar may not pick up small boats in bad weather or low visibility-the "Marine Check" will however, detect the other vessel's radar transmission and allow you to take avoiding action.

The scanning head is a miniature, precision, solidstate detector and amplifier of continuous, pulsed
and modulated X-band radar transmission ( 9,000 to $11,500 \mathrm{MHz}$ ) with electronics mounted in a shockproof case and will receive radar transmissions at sighting distance.

As the sensitivity is reduced, the directional power of the antenna is proportionately increased to a point where the bearing is located on a narrow sector. If this bearing, which is indicated on a compass scale, remains constant, you may be on a collision course, also as the pulse length increases as the distance decreases, it is possible, within limits to determine if the transmitting vessel is approaching or receding.

The "Seascan" radar by Electronic Laboratories (Marine) is designed for the amateur, and has a power consumption of only 48 W from $12 \mathrm{~V}, 24 \mathrm{~V}$ or 48 V battery supplies. The only vacuum devices used are the display tube and the magnetron, the local oscillator being a Gunn diode whilst the magnetron modulator uses a thyristor. Unfortunately no circuit details were available and all the units on the stand were dummies.

One interesting new development in radar for the small craft is a passive radar detector by Hepplewhite Marine. This picks up the X-band ( 9,000 to $11,500 \mathrm{MHz}$ ) transmissions from other vessels, and has a small directional aerial consisting of a Perspex "horn" on which waves are guided (along the surface) by the effect of the permittivity of the material. The received radar signals are mixed with the output of a local oscillator, and the beat note amplified. The point of strongest signal can be identified by altering the mixer bias so that only a signal of the maximum strength will produce a beat note; this enables the operator to find the bearing of the craft carrying the radar.

## SPEED MEASUREMENT

Water and wind speed instruments have for a long time been a weak point in the instrumentation of boats. The traditional methods of measuring speed have depended on anemometers for wind speed and submersible "logs" for water speed. The anemometer is the device using three or four cups at the ends of radial rods pivoted so that the effect of wind is to revolve the assembly about its axis. The


The Electra Magnetic Log is an impellerless speed and distance indicator.


Fig. 1: The Doppler Principle. Imagine a transducer giving out sound waves. These waves travel in water at a speed of about 1,000 metres per second. Speed equals frequency multiplied by wavelength.
If the transducer moves at a speed "V", the speed of the waves towards the reflector is now $1,000+V$ metres per second. We have not changed the frequency of transmission, so the wavelength must now be less.
$(1,000+V=f \times$ new wavelength. $)$
The waves are now reflected to a moving object, so that more waves per second hit the receiving transducer. The frequency is now higher by a factor $\frac{c+V}{c}$ where $c$ is

the normal wave speed and $V$ is the speed of the transducer and boat. The final effect is to receive waves of shorter wavelength and higher frequency.
speed of rotation, which depends on the wind speed, is measured by counting the number of switching operations of a reed switch under the influence of a magnet on the rotating shaft. This count is converted into an analogue signal which drives a meter. Wind direction is found by using a vane coupled to a potentiometer to show the angular position by voltage detection on a meter.

The traditional submersible "log" consists of a small propeller whose shaft is held in a casing containing a reed switch. The rotation of the propeller as the "log" is towed behind then registers relative water speed just as the anemometer registers relative wind speed. The snag is that the propeller of the $\log$ is easily jammed by anything encountered in the water from jellyfish to seaweed, and this can have the effect of making calculation of speed and distance incorrect.

## SPEED-LOGS

Two novel speed-logs were shown, each representing refreshingly different thinking about the problem. Space Age Electronics showed their Doppler speed log. This uses a ceramic transducer to project a 1 MHz sound beam ahead of the ship. Most of the beam is dissipated in the water, but a small part is reflected from the layer of relatively still water (the boundary layer) close to the boat.

When the boat is moving, the reflected frequency is not the same as the transmitted frequency (Doppler Effect). If the boat is moving ahead, the reflected frequency is higher than the transmitted frequency; if the boat moves astern, the reflected frequency is lower.

Referring to Fig. 1, the transmitter, because of its movement, "packs in" more waves between itself and the still layer than it would if it were not moving. Because more waves are travelling, the wavelength seems shorter. Also, when the waves reflect from the boundary layer, the receiver has also moved towards them, making the frequency of the received waves seem higher.

When the reflected wave is detected, the shorter wavelength appears as a higher frequency (speed $=$ frequency $\times$ wavelength), whose value depends on the difference between the speed of sound in water and the speed of the boat. By beating the transmitted frequency against the reflected frequency, a measure of speed can be obtained.
The method is dependent on obtaining a reflection from relatively still water, and so the transmitting/


Fig. 2: An e.m.f, is produced between two metal contacts, providing the direction of flow is at right-angles to the line joining the contacts.
receiving head must be located where such a layer is within range. Precise location varies from boat to boat. To allow for variations in installation, the calibration is adjustable, so that the distance log can be checked against accurately known distances (between two land-marks, for example). The circuitry uses three integrated circuits, 17 transistors and eight diodes. A "log" accuracy of better than $\pm$ $11_{2} \%$ (typical) is claimed.

The "Electra" magnetic log, produced by E.M.I. marine uses the magnetohydrodynamic principle for its operation; this is the generation of a voltage by induction when a conducting liquid flows past a magnetic field.

By arranging a magnetic field to cut the direction of water flow at right angles, an e.m.f. is induced between two metal contacts in the moving water, provided that the direction of flow is at right angles to the line joining the contacts (Fig. 2)

The Electra Magnetic Log operates on this principle, but with some ingenious modifications. The magnetic field used is an alternating field. In this way, the voltage generated is also alternating, which disposes of two problems at once. One problem is polarisation; when a steady voltage exists between two contacts in sea water, an electrolysis action takes place causing corrosion of the metal. The use of a.c. avoids polarisation problems and makes it easy to amplify and display the output from the contacts. As with the previous method, the positioning of the measuring head is critical.

## ECHOSOUNDERS

Echosounders have been used for a considerable time, and changes have been in detail rather than in concept. Basically, a marine echosounder consists of a pulse generator working at ultrasonic frequen'cies in the region 20 kHz to 150 kHz (the higher frequencies are used for inshore work). A transducer converts the electrical pulses into sound waves directed from the bottom of the boat into the sea, and a receiver, ising the same transducer, detects the pulse reflected from the bottom of the sea.

The shape of the received pulse provides a guide to the nature of the sea bed, for example, a muddy

sea bed is indicated by a broad pulse, a sandy bed by a narrow pulse, a rocky bed gives multiple reflections. Pulses received from intermediate depths can indicate shoals of fish. The time delay between transmission and reception is proportional to the depth of sea.

Because the pulses are travelling in the water at the speed of sound (around 1,000 metres per second) rather than at the speed of radar waves in space (some 30 million metres per second), scanning rates have to be lower than the $1,000 \mathrm{~Hz}$ commonly used in radar. Scanning rates of once per second are common, although this is rather too low for display on a long persistence cathode ray tube screen.

The display method traditionally used is a rotating wheel carrying, or illuminated by, a neon which flashes on transmission of the pulse and again when the echo is received. The stationary bezel round the wheel is then calibrated in metres of depth (now replacing fathoms on the latest charts).

The large number of echo sounders at the show demonstrated several novel points of design. Most interesting was the use of electro-luminescent diodes to replace neons, so avoiding the need to generate a high voltage supply (Brookes \& Gatehouse). A common feature of all the echo sounders was the use of high scan rates for the shallow water ranges (up to six pulses per second) so giving better definition in critical conditions.

Lead zirconate titanate crystals were in evidence as transducers, encapsulated in various plastics. For professional use, chart recording echosounders were shown (these, of course, have been around for a long time). The Marconi Marine chart-maker used electrosensitive paper which changed colour when a small voltage was applied between the front of the paper and the back.

## WATER-LEAK AND WATCH

In addition to the displays of electronics for communication and navigation, there were some interesting exhibits in the "miscellaneous" class. SpaceAge Electronics were showing their water-leak and watch alarm, and judging by the sound which could be heard all over the first floor of Earls Court, almost every chandlery firm was demonstrating it.

The instrument is basically a conventional leakage detector using stainless steel pins as probes. If the pins are immersed in water for more than four seconds, an alarm is given, but only if the immersion is for longer than four seconds, a time carefully chosen to avoid continual false alarms in heavy seas. When there is no conductivity between the pins, all transistors are biased off, so ensuring practically no standby current. The sound given out is a rising frequency note which is most easily distinguished amongst other sounds.
The watch facility also provided on this instrument gives an alarm note of gradually increasing amplitude every three minutes. If an alarm button is pressed during the alarm, or before it, the instrument remains silent until another three minutes has passed. If the helmsman is asleep and does not press the alarm button, the alarm warning increases to maximum intensity.

## GAS ALARM

The Electra gas alarm, from E.M.I. Marine, is of considerable interest. Unlike previous gas alarms, which used heat conduction from a hot wire to detect changes in the air composition, the Electra gas alarm uses semiconductor sensors. No details were given, but it is possible that the sensors consisted of infra-red electroluminescent diodes and detectors. Any change in the air composition caused by petrol vapour, butane or propane gas, carbon monoxide, smoke, paraffin fumes, diesel fumes is indicated by two visual warnings in the form of a meter reading and a red warning lamp.

In addition, a loudspeaker gives out a 2.5 kHz note when the gas concentration exceeds the safe limit. In this way, no panic is caused by small concentrations of gases, since only the discreet visual warning is given, but dangerous amounts operate the audible alarm so that evacuation or other measures can be taken. Remote warnings can also be fitted.

All in all, the Boat Show is developing considerable interest from the point of view of electronic aids. It is unfortunate that so few exhibitors were able to cope with technical queries, but all of them were most helpful.

# practically Wireless HENRY 

IHAD a chap in the shop today, searching for an SP4. It is true: even in the glitter of a hi-fi establishment, where every prospect teases and only valves are vile, we still get enthusiasts with thirty-year-old equipment, hopeful of our being able to provide the missing link.

Which leads us, philosophically, to a contemplation of some of the 'firsts' we nowadays take for granted.

Example: the quasi-complementary transistor amplifier circuit, which has given us so much trouble since its inception. I have come across the September 1956 copy of Electronics, a McGrawHill publication, in which R. C. Lin, then of RCA, Princeton, first propounded his output circuit in which, to quote: '. . . the two upper transistors conduct during the negative half-cycle and the two lower transistors conduct during the positive half-cycle.'

In his summary, he gave amplifier distortion at 100 cps (sic) and 400 cps as below 1 per cent at six watts, and though he admitted that the distortion might be expected to rise at $5,000 \mathrm{cps}$, he made no mention whatsoever of its fierce spikiness at a fraction of a watt, which is the mariner we've had wrapped around our necks ever since.


The mariner wrapped around our necks.

Inventions, especially worldshaking innovations such as the transistor proved to be, have a habit of spawning themselves much more prosaically than we would like to imagine. No technologist runs screaming through the corridors of the M.I.T. with 'Eureka' echoing behind him. More likely, something as laconic as the $12 / 29 / 39$ : 4.15 p.m. entry in William Shockley's notebook may be handed down to posterity: 'It has today occurred to me that an amplifier using semiconductors rather than vacuum tubes is in principle possible.

In principle: hardly in practice. He was thinking of a Schottky-barrier FET kind of device, with copper-oxide as the semiconductor. And, as Henry could have told him*, it didn't work. Instead, the experiments of the team included covering metal points with wax and pushing them down on to p-type silicon treated to give an n-type surface. They surrounded the point with water and found they could obtain power amplification.

To quote Brattain: '. . . the group was jubilant that day.'

Trouble was that the water would evaporate. Maybe it was all that heavy breathing. So they switched to glycol borate to reduce evaporation. One trouble: amplification was obtainable only below 8 Hz -hardly hi-fi yet.

There were lots more experiments, culminating in evaporated gold spots and germanium, and then, on Dec. 23rd 1947, two gold contacts less than two-thousandths of an inch apart were made to the same piece of germanium and the first transistor happened.

No sudden breakthrough-no inspirational brainchild-and no immediate christening, I may add. It was a month later that a chap called John R. Pierce of the Bell technical staff mooched into Brattain's office and pointed out that the 20 dB point-contact amplifier they had made was the


Runs screaming through the corridors.
'. . . dual of a vacuum-tube, circuit-wise.' He mentioned the important valve parameter, transconductance. Then he went on to talk about its electrical dual, transresistance. To quote Brattain once more: '. . . then he said 'transistor', and I said, 'Pierce, that is it!!'

What about a few more firsts in our field? Who developed the first p-n junction? Bet you can't even say his name, let alone remember it. Well Henry can, with the benefit of some help from Edward A. Torrero, who is Associate Editor of Electronic Design. He was William Pfann (with the aid of a chap called Jack Schaff; let's give credit where it is due). They also worked for Bell Laboratories.

But Bell eschewed silicon for germanium, because the impurities, back in the early 1940's, were easier to control. It was left to Texas Instruments to develop silicon types of transistor, and to Motorola to introduce the diffused base transistor.

So I could continue. Henry hasn't the space to name those to blame.

But the "first" we are all concerned with is that original alltransistor radio. And who made that? A firm called Regency, in 1954. How's that for a significant 'first'?

* With the benefit of hindsight. Ed.


# EXTRA \#\# INNEXTMONTH'S <br> PRAGTIGAL miallas <br> <br> (组 <br> <br> (组 GPAGESUPRLEMENT 

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P.W's reputation for plenty of simple and straightforward constructional projects is excelled next month by an extra large issue. In addition to the usual constructional articles and features, the May issue will include a special extra 8-page coloured supplement describing eight easy projects for the home, designed and described by one of our top designers; you are bound to find at least one of these of interest. This supplement is fully illustrated with all the details on how to build these projects.



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Following the success of the Texan amplifier published in this magazine during May to August 1972, the applications laboratory of Texas Instruments Ltd have developed exclusively for $P . W$, a compatible unit for the x frominol of Psychedelic lights from a sound isource. The initial design features zero voltage 3xopifolta A more sophisticated design incor poratins lamp dimming facilities will also be covered th this series of articles For stered $0 p$ thation tho $P W$. Tricolour units are reQurer

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## MONO/STEREO CAP ABILITY

ZERO VOLTAGESWITCHNG

## NEGLIGIBLE RADIO INTERFERENCE

## LAMP OVER-RIDE CONTROL

OPTIONAL LAMP DIMMING

COLOUR
ference (r.f.i.) due to the step change in current. The noise thus generated is spread by means of conduction through the wiring system or by radiation. Suppressing r.f.i. becomes more difficult and expensive as the load increases. Uncontrolled switching of "psychedelic" lamps on and off will have a similar effect and will also introduce additional noise in the loudspeakers. A simple solution to this problem is to arrange the switching so that it always takes place at the point where the mains supply voltage crosses zero, as shown in Fig. 3.


Fig. 3: Waveform depicting zero voltage switching.

This technique is called 'zero voltage switching' or 'burst firing.' In this way isolated numbers of half or full cycles will pass through the lamps and thus avoid step change in currents, which also has the additional advantage of increasing the life of the lamp.


Fig. 4: Basic circuit which generates the pulses at the point of zero voltage crossing.

Fig. 4 shows the circuit used to generate the pulses at the point of zero voltage crossing. Transistor Trl, whose base is connected to the unsmoothed supply from the bridge rectifier D1 is on most of the time.

However, each time the mains voltage crosses zero, its base voltage is taken below the holding on voltage and it thus turns off, giving positive pulses on its collector. In order to reduce the loading on transistor $\operatorname{Tr} 1$, transistor $\operatorname{Tr} 2$ has been added. Positive pulses appear as an output across resistor R each time the mains voltage crosses the zero point.


Fig. 5: Typical music voltage waveform at the terminals of a loudspeaker.

In order to control the lamp switching with respect to the frequency bands, we need a circuit which will separate out the low, middle and high frequencies from the voltage waveform appearing at the terminals of a loudspeaker. A typical voltage waveform appearing at this point is shown in Fig. 5.

## Filter

A circuit which separates frequencies into bands is called a filter. There are various types of filters which could be employed; e.g. one commonly used is a resistor/capacitor (r.c.) network for high and low pass filters and an inductor/capacitor (l.c.) network for middle frequencies. The elegant way, however, specially developed for the P.W. Tricolour is to employ integrated circuit (i.c.) operational amplifiers, in a positive fixed-gain configuration, as the active filter element and resistors and capacitors as passive elements. By choosing suitable component values a sharp separation between bass, middle and treble frequencies can be obtained.

The shape of the filter response is determined by a factor known as the damping ratio ( () . The lower this ratio, the sharper the cut-off and there is a peaking effect. This peaking effect is of no importance in our application. As approximately 1.5 volts is required to turn on the interface circuit, and thus to switch on the appropriate lamps, the filter characteristics are expressed in volts against frequency and not in the more usual decibels against frequency.


Fig. 6: Basic low pass filter as used in the P.W. Tricolour.

The theory of filter design is quite extensive and complex. We will however only outline the necessary basic concepts and give some simplified formulae.

Let us first look at a basic low pass filter (i.e. one which will stop all frequencies except those below a chosen frequency). The circuit of such a filter is shown in Fig. 6.

Here the angular cut off frequency,

$$
\omega 0=\frac{1}{\mathrm{R} \sqrt{\mathrm{ClC} \mathrm{C}}}
$$

and the damping factor

Critical damping occurs when

$$
\begin{aligned}
\zeta & =\sqrt{\frac{\mathrm{C} 2}{\mathrm{Cl}}} \\
\zeta & =\frac{1}{\sqrt{2}}
\end{aligned}
$$

and, for the reason explained before, i.e. to improve the sharpness, this value is reduced to $0 \cdot 5$. From these equations, the filter component values can be established.
The high pass filter (passing all frequencies above a chosen one) is shown in Fig. 7 and again requires only one operational amplifier, two resistors and two capacitors.

Here the angular cut-off frequency,

$$
\begin{aligned}
\omega 0 & =\frac{1}{\mathrm{C} \sqrt{\mathrm{R} 1 \mathrm{R} 2}} \\
\zeta & =\sqrt{\frac{\mathrm{R} 1}{\mathrm{R} 2}}
\end{aligned}
$$

Fig. 7. Basic high pass filter.

TE79

Fig. 8: Basic band pass filter.
Once more from these equations, the filter components values can be established. Finally, the band pass filter; which, of course, as its name suggests, passes a band of frequencies. This, as can be seen from Fig. 8, is a little more complicated than the other two.

The centre angular frequency

$$
\begin{aligned}
\omega 0 & =\frac{\sqrt{2}}{\mathrm{RC}} \\
\mathrm{Q} & =\frac{\sqrt{2}}{4-\frac{\mathrm{R} 2}{\mathrm{R} 1}}
\end{aligned}
$$

$\mathrm{Q}=\frac{2 \Delta \mathrm{f}}{\mathrm{fo}}$, where $2 \Delta \mathrm{f}$ is the difference between two frequencies at which the voltage output of the filter is $\frac{1}{\sqrt{2}}$ of its maximum value. From the above formulae, relevant component values can be obtained. ormulae, relevant component values can be obtained.
Musically oriented readers will recall the 'unisono' note played on the oboe during the orchestra's tuning-up time, just before the concert begins. This note is middle ' $A$ ' with a frequency of 440 Hz ; a frequency which has been chosen as the centre frequency for the middle band-pass filter in


The centre angul fror
and the quality factor
$\qquad$ two frequence at which the voltage output of the - .

Rear view of the P.W. Tricolour.



Fig. 9: Complete P.W. Tricolour filter circuit. This unit is used without modification, with either the zero voltage switching unit (Fig. 12) or with the lamp dimming unit, to be described later in this series.
this design. Cut-off frequency for the low pass filter has been chosen to be around 200 Hz and that of the high pass filter (treble) about 1 kHz .

Based on the above assumptions, a complete filter has been developed with the component values as shown in Fig. 9. The alert reader will notice that some of the component values differ from the figures obtained using the formulae. This has been done intentionally in order to improve the separation between the various frequency ranges.

The input to the filter circuit has been limited to 5 volts peak to peak by using two 5 volt zener diodes connected back to back. Three i.c. operational amplifiers are used. Two of which are contained in one 8 pin package, SN 72558 P and the other i.c. is contained in a second 8 pin package, SN72741P.

Fig. 10 shows output voltage response of the three filters plotted against frequency for a peak input signal of 0.5 volts. The effectiveness of the filters are obvious. The solid state switches-triacs-turn on their corresponding lamps at filter output voltages equal to and higher than 1.5 volts.

Figs. 11a to Ild show four oscilloscope photographs. These show traces with the vertical axes representing voltages and the horizontal one time. In each instance the four traces were taken simultaneously (i.e. they have a common time axes). In the oscillograms the top trace corresponds to the voltage at the loudspeaker terminal. The other three traces were taken at the outputs of the respective filters.

Fig. Jla was taken at a time when the treble


Fig. 10: Output voltage/frequency response of the filter circuit shown in Fig. 9.


Fig. 11 : ( $a-a^{i}$ ) A full explanation of these waveforms is given in the accompanying text.
frequency was predominant. In Fig. 11b the middle frequency is the strongest signal, while in Fig. 11c it is the bass. Fig. 11d shows some interesting combinations of all three frequencies. The middle frequencies predominate at the beginning of the trace, then the treble takes over at the centre of oscillogram and towards the end of the trace it is the turn of the bass to predominate.

A $10 \mathrm{k} \Omega$ potentiometer has been added at the input to the filter circuit to allow the voltage level at which the lamps are switched to be controlled. In this way it is possible to establish the best effect for 'soft' or 'party' type music.

The P.W. Tricolour is suitable for use with 4 to $15 \Omega$ loudspeakers. The load represented by the filters is negligible compared to the loudspeaker impedance.

It should be noted that the simple filter circuits illustrated at the beginning of this article used operational amplifiers. An operational amplifier has high input impedance, very high gain, and low output impedance, enabling the filter component values to be determined independent of the amplifier parameters. The amplifiers used were shown having a voltage gain of unity. This means that the output voltage of the high and low pass filters can never


Underside view of the P.W. Tricolour with zero voltage switching.


Fig. 12: Zero voltage switching control unth. The positive and negative 10 V supply lines are the d.c. supplies to the filter unit (FIg. 9).
be more than the input voltage. In order to achieve voltage gain so that sufficient output can be obtained to drive the interface circuit at low volume levels, the basic circuit must be modified. This involves reducing the proportion of output voltage from the amplifier that is used as feedback to the filter capacitors and resistors. This is achieved by the use of a resistive potential divider at the output of the
amplifier. This is shown in Fig. 9. The resistor ratio determines the gain of the filter in the pass-band. The P.W. Tricolour is suitable for either mono or stereo record playing. In the case of the stereo system, the lights will flash in accordance with the signals appearing at the terminals of one of the speakers. However, for stereo enthusiasts who wish to use two units, the interconnecting arrangement will

## * components list



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## TRANSISTORS BY SIEMENS AND NEWMARKET <br> 2N3055 npit silicon power 60p AClis3K pnp germanium low power 25p AC1F6K npn germasium low power 23p AD1f1 npn gertantum ntedium power 42p AD162 pnp germanium mediunt power 40p AF139 phy germanium UHF 39p $\left.\begin{array}{l}\text { BCN 10: 13p; BC108 12p; BC109 13p; } \\ \text { BC167 10p; BC1 }\end{array}\right\}$ 日p BC17 15p; BC178 14p; BC179 15p; f pno BC257 9p; BC258 8p; BC259 9p <br> Etandard groupings available. <br> BDI35 npa med puwer 28p <br> BD136 pap med power 27p <br> DIODES <br> OA90, OA91, OA95 each 6p <br> OAz00 日p; OA202 10p <br> Other semiconductors AC128 21p; AF117 24p <br> BFY51 20p <br> Full lists and technical data will be found in Catologue No. 6. <br> SIEMENS' <br> THYRISTORS

$0 \cdot 8 \mathrm{~A} 400 \mathrm{~V} 56 \mathrm{p}, 600 \mathrm{~V} 70 \mathrm{p}$
$3 \mathrm{~A} \quad 400 \mathrm{~V} 60 \mathrm{p}, 600 \mathrm{~V} 88 \mathrm{p}$
ZENER DIODES full range E24 values: 400 mW ; 2.7 T to $36 \mathrm{~V}, 14 \mathrm{p}$ each; $1 \mathrm{w}: 6 \cdot 8 \mathrm{~V}$ to 82V, 2 pp each; $3.6 \mathrm{~W}: 4.3 \mathrm{~V}$ to $75 \mathrm{~V}, 48 \mathrm{p}$ each. Clip to inerease l -5W rating to 3 watt (type 266 F ) 4 p .

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2 way Ls-socket 10p; pug 12p. 3 way scr. socket 10p; plug 12p. 5 way ser. socket 12p; plug 15p.

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Heat sinks $1^{\circ} \mathrm{o} / w^{\prime}$, type 6W1. undrilled 60 p
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1011 C SPST 409 DPDT ${ }^{19 p}$ toggle 28p. (These a
25 A rating) j201 Subrininiatur 250V a.c/2A 48p ROTARY SWITCHES
Radiospares Miniature Maka-suitcte (in assembly kit form) Shaft 48p. Waiers, MBB-2PSW, JPIIW: 3BM-1P12W, 2P6W $3 \mathrm{P} 4 \mathrm{~W}, 4 \mathrm{P} 8 \mathrm{~W}, 6 \mathrm{~F} 2 \mathrm{~W}$. each, 6 p .

WAVECHANGE SWITCHES

1P12W, 2PifW, 3P4W 4 P3W, each 24p
axial lead
$\begin{array}{lllllllll}\text { Rated voltage: } 3 \mathrm{~V} & 6.3 \mathrm{~V} & 10 \mathrm{~V} & 18 \mathrm{~V} & 25 \mathrm{~V} & 35 \mathrm{~V} & 50 \mathrm{~V} & 83 \mathrm{~V} & 100 \mathrm{~V}\end{array}$
Capacity $\mu \mathrm{F}$
0.47
1.0
2.2
4.7


RESISTORS 10\%, 5\%, 2\%

| Code | Poteer | Tolerance | Range | Values | 169 | 10 to 99 | 100 up |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | available | (see note below) |  |  |
| 0 | 1/20w | $5 \%$ | $82 \Omega-220 \mathrm{~K} \Omega$ | E12 | 0 | 8 | 7 |
| C | 1/8W | 5\% | $4.7 \Omega-470 \mathrm{~K} \Omega$ | E24 | 1 | 0.8 | $0 \cdot 7$ |
| C | 1/4W | $10 \%$ | $4.7 \Omega-10 \mathrm{M} \Omega$ | E12 | 1 | 0.8 | 0.7 |
| 0 | 1/2W | 5\% | $4 \cdot 7 \Omega-10 \mathrm{M} \Omega$ | E24 | $1-2$ | 1 | 0.9 |
| c | 1W | 10\% | $4 \cdot 7 \Omega-10 \mathrm{M} \Omega$ | E12 | 25 | 2 | 1.9 |
| MO | 1/2W | 2\% | $10 \Omega 1 \mathrm{M} \Omega$ | E24 | 4 | 3 | 2 nett |
| WW | 1W | 10\% $\pm 1 / 20 \Omega$ | $0 \cdot 22 \Omega-3.9 \Omega$ | E12 | 7 | 7 | 6 |
| WW | 3W- | $5 \%$ | $1 \Omega-10 \mathrm{~K} \Omega$ | E12 | 7 | 7 | 6 |
| Ww | 7W | $5 \%$ | $1 \Omega-10 \mathrm{~K} \Omega$ | E12 | 0 | 9 | 8 |

Polycarbonate $-5 \%$ Tolerance 250V up to $0.1 \mu \mathrm{~F}: 100 \mathrm{~V} / 0 \cdot 1 \mu \mathrm{~F}$ and $\begin{array}{llll}\text { above. } \\ 0.01 & 012 ; & 0.015 ; & 0-018 ; \\ 0.022\end{array}$ $0.027 ; 0.033 ; 0.047 ; 0.056$; each 4 p . 0.068; $0.082 ; 0.1 ; 0.12 ; 0.15$ each 4p. 0.18; ;2.2; each 5p. $0.27 ; 0.33: 6 p .0 .397 p .0 .478 p .0 .5610 p$
0.6811 p .14 F 0.68 11p. $1 \mu \mathrm{~F}$ 13p.

C-cambon fllm, high stability, low noise. Mo--metal oxide, Electrosil

Prices are in pence each for quantities of the same ohmic value and power rating. NoT mized valnes. (Ignore fractions on total value ol resistor order)

Values:
E12 denotes series: $10,12,15,18,22,27,33,39.47,56,68,82$ nind their decarles.
E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 81 .
$62,75,91$ and their tecades.
62, 75, 91 and their decades.
EVERYTHING BRAND NEW, GUARANTEED TO MAKERS SPEC. NO SECONDS OR REJEGTS


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## Tone control circuitry

Toward the end of the last part of this series of articles that was published, we were concerned with frequency-dependent feedback. Without being intentionally devious, we skated over some of the finer points and here shall try to keep our promise of digging more deeply into the matter of tone controls, their design and their limitations.

The utter purist will not deign to acknowledge the need for tone controls! Everything should be flat, he says. There should be, he further asserts, no need to alter the frequency response of an amplifier if all the other parts of the audio system are in order. True, but Utopia has not yet arrived, whatever the politicians promise, and there will be nasty pops and crackles from our records, hiss from our tape, peaks and troughs from many a cartridge and even worse deviations by the time the sound escapes from the loudspeakers.

Even then, we are not finished. If the little woman insists on foot-deep rugs and wall-to-wall drapes, or if your pad is scoured to its functional austerity, man, you will need same trimming of the ideally flat response to compensate for-in the first case, the premature absorption of treble, and in the second, the 'hardness' of the sound. Tone controls may be regarded as necessary evil, but as long as we are obliged to have them, let's make sure they do not detract from the best possible performance. And in so many commercial designs, they do.

## First steps

Before delving into the tone control circuit as a whole, we must take a look at the effect of the passive components with a change in frequency. Back in the days when Poppa bought his radiogram because it sounded 'mellow', the tone control was simply a network of passive components, usually across the output stage, with variation of the resistive part of a series capacitor-resistor shunt that drained away progressively more and more of the upper frequencies. It was, in other words, a variable treblecut device. One or two more enterprising makers


Fig. 74: (a) For a simple resistive network the output shown by the voltmeter is in proportion of the ratio of the resistors to the applied input voltage. In (b) theoretically, the output should not change with frequency, giving a "straight-/ine" graph. (c) Substituting a capacitor for R2 makes the output frequency dependent while (d) shows an example of the curve we might expect from a simple network such as (c) having a "slope" of $6 d B$ per octave.
fitted alternative resistor-capacitor combinations in the grid circuit of output valves, and others even went so far as to label these Bass or Treble Boost and Cut, but all such passive networks, i.e., those which do not have inbuilt amplification, must attenuate the overall signal.
The point about the passive tone controls is that the attenuation varies the frequency. Take the case of a simple resistive network, as in Fig. 74(a). If we have two resistors of equal value, fed by a signal generator, the output across the measured one will be half the input signal in amplitude, but will not change in frequency. A graph of the situation would be a straight line, as Fig. 74(b).
But if we now substitute the upper resistor with a capacitor, Cl in Fig. 74(c), we find that the measured output will be exactly half the input at only one frequency. Reactance $X_{c}=\frac{1}{2 \pi \mathrm{fC}}$ where $\mathrm{X}_{\text {. }}$ is in ohms, C in farads and f is the frequency at which the test is made, in hertz.
Working out the reactance of a $l \mu \mathbf{F}$ capacitor at 1 kHz , we get $\frac{10^{6}}{6.28 \times 1000 \times 1}=159$ ohms.

But if we now measure the voltage across a similarvalue resistor, we find that instead of half the signal input, we have slightly more. Instead of our attenuator dropping the output by 6 dB (to half voltage) at 1 kHz , it will have dropped the output only 3 dB . This is because the voltage across the capacitor is $90^{\circ}$ out of phase with the voltage across the resistor. The output voltage ratio is $0 \cdot 707: 1$, not $0 \cdot 5: 1$. In fact, if we alter the frequency of the generator, we will find that the measured output drops as the frequency falls. The rate of 'slope', as shown in Fig. 74(d), is 6 dB per octave, an octave being a frequency change of $2: 1$.

This 3 dB drop has a further significance when we tackle power output stages, for if the drive voltage to the amplifier fell sufficiently for the power output to drop 3dB, this would be the well-known 'half-power-point'. This is the figure normally quoted in amplifier specifications-a point to which we shall return later. Below this -3 dB point in Fig. 74(d), the slope of the curve falls away regularly by 6 dB per octave, i.e., a ratio of $2: 1$, a reduction of output by a half each time the frequency is halved.

## Making it active

It helps to regard tone control as a method of frequency correction. The introductory remarks we have made, and references to frequency-dependent


Fig. 75: A collector-follower amplifier arranged so that its output can be measured with and without feedback.


Fig. 76: The basic circuit used for tests.
feed-back in previous articles, should have prepared the regular reader for this approach. Simple tone controls are types of filters. We shall deal thoroughly with filter circuitry later, but at the moment it is necessary to observe that the 'filter' of Fig. 74(c), for example, is called a 'first order filter', having a 'slope' of 6 dB per octave. Second order filters have slopes of $12 \mathrm{~dB} /$ octave and third-order filters, $18 \mathrm{~dB} /$ octave. As we shall see, real filters, used in welldesigned audio amplifiers, should ideally have a third-order slope to be effective.
These filters are passive devices. They contain no amplifying stage or stages to put back the signal attenuated by their action. If we make a marriage of the virtual earth circuitry we talked about in Parts 12 and 13 of this series and the fundamentals of tone control circuitry touched upon in Part 11, we get an amplifier with frequency-dependent feedback, which becomes, if we correctly design it, a better tone control circuit.

First, we need what is known as an 'inverting amplifier'. If you take a look at Fig. 75 , you will see that this can be a collector-follower circuit, so arranged that we can measure its gain with or without the feedback. In practice, this is a useful circuit, and the component calculation will be discussed as we go along. It is convenient to use the 24 -volt supply described in Part 13, December PW. Fig. 76 shows the circuit devised by Mike and used for tests on the prototype.

With an input from the audio generator, and an output measured with a millivoltmeter, we can measure the open-loop gain. This is the gain of the amplifier with no feedback connected. Here, a word of warning is required: in many erudite theoretical articles, 'open-loop gain' is bandied about freely, but an inspection of the circuit will often show that it would be virtually impossible to measure it in practice. Our aim in this case is to show what we mean by the term, then, by applying the feedback we need, to modify the gain and achieve our controlled response, i.e., effect a tone control.

In our prototype, the open-loop gain, ( $\mathrm{G}_{0}$ ) was 260. It is absolutely essential when carrying out these measurements to mount the circuit in a screened box, use screened input and output leads and make sure our construction is 'solid'. We use Veroboard for convenience: it allows connection of additional components while retaining the versatility of wired circuitry. We use the very convenient Norman Rose aluminium boxes, because they are inexpensive, easily worked for outlet holes, tag-strip mountings, etc., and have a simple lid which not only completes the screening but also affords an easy access for modifications.

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| 1 mat | 22.15 | 50Y. D.C. .. 52.15 |
| $\stackrel{3 \mathrm{ma}}{10 \mathrm{ma}}$ | ${ }_{52}$ | 300V. D.C. $\quad \frac{52.15}{82.30}$ |
| 50 ma | E2.15 |  |
| 100ma | \&2.15 | 300 V . A.C. . . 22.30 |
| 500 ma | £2.15 | VU Meter .. $£ 2.55$ |

*MOVING IRON-
all others moving coil Please add postage

| - |  |
| :---: | :---: |
|  | 100 mA .... $£ 3.10$ |
|  | 500 mA . ... $£ 3.10$ |
|  | 1 amp....... $\$ 3.10$ |
|  |  |
|  | $\begin{array}{llll}15 \mathrm{amp} & . . . & £ 3.10 \\ 30 \mathrm{mmp} & . . . & £ 3.90\end{array}$ |
|  | ${ }^{20 \mathrm{~V}, \text { D.C. }}{ }^{\text {c }}$ + ${ }^{2} .10$ |
|  | $50 \mathrm{~V} . \mathrm{D.C}$. . 88.10 |
| $50 \mu \mathrm{~A}$. . . . $£ 3.05$ | 150V. D.C. 88.10 |
| $50-0-50 \mu \mathrm{~A}$ - 58.40 | 300V. D.C. 53.10 |
| $100 \mu \mathrm{~A}$. . . 88.40 | 15Y. A.C. . 23.10 |
| $100-0-100 \mu \mathrm{~A}$ - 2380 | 300 V . A.C. .. $\$ 3.10$ |
|  | S Meter lina 53.15 |
| $500 \mu \mathrm{~A} \quad . .$. | VU Meter |
| 500-0-500 $\mu \mathrm{A}$ ¢3-10 | 1 amp A:C.* ${ }^{\text {che }} 10$ |
|  | 5 amp A.C.* 83.10 |
| 1-0-1 nA A .. 83.10 | 10 amp . A.C.* ${ }^{23} 10$ |
| 5mA ...... 83.10 | 20 aimp . A.E.* 88.10 |
| 10 mA . . . 83.10 | 30 amp . A.C.* 28.10 |
| Type MR.52P. 2tin. square fronts |  |
| $50 \mu \mathrm{~A}$...... 82.40 | 10V. D.C. . 22.20 |
| $50-0-50 \mu \mathrm{~A}$ 22-85 |  |
| $100 \mu \mathrm{~A}$... 28.85 | 50 V. D.C. . . 52.20 |
|  | 300 V . D.C. $22-20$ |
|  | 15V. A.C. . . 82.30 |
| $1 \mathrm{~mA} . . .$. | 300 V . A.C. . 2230 |
| $5 \mathrm{maA} . . . .{ }^{22-20}$ | S Meter Ima |
| $10 \mathrm{~mA} \mathrm{...}$. | YU Meter . ${ }^{\text {a }}$ E 50 |
|  | 1 amp. A.C. * 82.20 |
| 100 mA .... 28.20 | 5 amp A.C. ${ }^{*}$ ¢ $£ 2-20$ |
| 500 mA .... 22.20 | 10 amp. A.C. ${ }^{\text {ced } 20}$ |
| 1 amp . . . . . ${ }^{\text {e2.20 }}$ | 20 arap. A.C. * 28.20 |
| $5 \operatorname{amp}$. . . . . 28.20 | 30 amp. A.C. ${ }^{\text {P }} \mathbf{8 2} \mathbf{2 0}$ |
| Type MR,65P. 3itin. x 3 ${ }^{\frac{1}{2} \text { in. fronts }}$ |  |
| 50 $\mu \mathrm{A}$. . . . . $23 \cdot 70$ | 10V. U.C. . 2240 |
| 50-0-50 A A 88.00 | 20V. D.C. . 5240 |
| $100 \mu \mathrm{~A}$. ... 28.00 | $50 \mathrm{~V} . \mathrm{D} . \mathrm{C} . \operatorname{CS} 90$ |
| 100-0-100 1 A 22.90 | 150 V . D.t. 82.40 |
| $200 \mu \mathrm{~A} \quad . . .8290$ | 300Y. D.C. $£ 2.40$ |
| $500 \mu \mathrm{~A} \quad \cdots \cdots$. 28.65 | 15V, A.U. .. $£ 255$ |
| 500-0-500 $\mu$ A 28.40 |  |
|  | 150V. A.C. . $52-55$ |
| $5 \mathrm{~mA} . . . . .$. | 300 V . A.C. . 22.55 |
| 10mLA .... $22 \cdot 40$ | 500v. A.C. 28.55 |
|  | S Meter Imía $£ 260$ |
| 100 mA .... 22.40 | VU Meter - $88 \cdot 70$ |
| $500 \mathrm{~mA} . . .9$ ¢ 28.40 | 50mA A.C. " ${ }^{\text {c2. }} 40$ |
| $1 \mathrm{amp} . . . . .$. . 28.40 | 100 mA A.O. * 28.40 |
| 5 amp....... . $82 \cdot 40$ | $200 \mathrm{~mA} \mathrm{A.C} *$. |
| 10 amp. . . . 82-40 | 500 mA A.C. * 88.40 |
| $15 \mathrm{mmp} . . . .$. | 1 amp . A.C. ${ }^{* 22} 40$ |
| $20 \mathrm{amp} . . . . .82 .40$ | ${ }^{5} \mathrm{amp}$ A.C. ${ }^{*} 22.40$ |
| $30 \mathrm{amp} . . . .585 .55$ | 10 amp . A.C. * 22.40 |
| $50 \mathrm{amp} . . . . .82 .75$ | 20 amp A.C. $* 28.40$ |
| 5V. D.C. . . $\mathbf{5 2} \mathbf{4 0}$ | 30 amp . A.C. * 82.40 |


"SEW" EDGWISE METERS Type PE.70. 3 17/32in. $\times 1 / 15 / 32 \mathrm{in} . x$

|  | 23 ${ }^{4}$ | eep |  |
| :---: | :---: | :---: | :---: |
| $50 \mu .4$ | E3.40 | $500 \mu$ | £3.05 |
| 50-0-50 5 /, | 43.30 | 1 mA | £2.7 |
| $100 \mu \mathrm{~A}$ | 83.30 | 300 |  |

 | $200 \mu \mathrm{~A}$ | $\cdots \cdots$ | $\$ 3.20$ | VU Meter | .. |
| :---: | :---: | :---: | :---: | :---: |
| 23.75 |  |  |  |  |

'SEW"' BAKELITE
PANEL METERS
Type MR.65. $3 \frac{1}{2}$ in. square fronts

|  | 1 amp....... 22.15 |
| :---: | :---: |
|  | 5 amy. . . . . . 22.16 |
| cied ${ }^{\text {a }}$ | 15 amp. . . 22.15 |
|  | 30 amp. .... $48 \cdot 16$ |
|  | $50 \mathrm{amp} . . .$. |
|  |  |
|  | 10V. D.C. . 88.15 |
|  | 20 V . D.C. . 28.15 |
|  | 50 V . D, C. . 28.15 |
|  | 150 V . D.C. 28.15 |
|  | 300 V , D.C. 22.15 |
|  | 50 mV . D.C. ${ }^{54.40}$ |
|  | 100 mV . D.C. 82.40 |
| $50-0-50 \mu \mathrm{~A} \quad 2 \mathrm{E}$ 20 | 30V. A.C. *. 4280 |
| $100 \mu \mathrm{~A}$ ( $\ldots$.. 22.60 | 50 V. A.C. ${ }^{\text {² }}$ : 42.20 |
| 100-0-100 2 A 2850 | 150 V. A.C. * 28.20 |
| $500 \mu \mathrm{~A} \quad \cdots \cdot{ }^{-9.45}$ | 300 V A.C. ${ }^{*}$ (22.20 |
| $500-0-500 \mu \mathrm{~A}$. 22.15 | 600mA A.C. * 22.15 |
|  | 1 amp. A.C. * 22.15 |
| 1~0-1mA .. 82.15 | 5 arup A.C.* 22.15 |
| $5 \mathrm{~mA} . . . .$. | 10 amp , A.C. * 22.15 |
| 10mA $\ldots$. 22.15 | 20 anıp. A.C. $* 88.15$ |
| 50 mA . 3. | 30 amp . A.C. * 22.15 |
| 100mA .... 28.15 | 50 amp A.C. ${ }^{*} \mathbf{8 2 . 1 5}$ |
| $500 \mathrm{~mA} . .$. | VU Meter . . 88.40 |
| Type S-80 80 m | square fronts |
| 50\%A .... : 53.50 |  |
| $50-0-50 \mu \mathrm{~A} \quad 93.40$ |  |
| $100 \mu \mathrm{~A}$. ${ }^{\text {c. }} 23.40$ |  |
| 100-0-10f $\mu \mathrm{A}$ A 53.30 |  |
| $500 \mu \mathrm{~A} \quad \cdots$. |  |
|  |  |
| 20V. D.C. . 82.85 |  |
| 50V. D.C. . . 22.85 | 5 anilu. V.C. 42.85 |
| 300 V . D.C. 22.85 | 300 V . A.C. . . 28.85 |
| 1 amp D.C. 82.85 | VU Meter . . 83\%70 |

## "SEW" EDUCATIONAL

Type ED.107. Size oyerall $100 \mathrm{~mm} x$ $90 \mathrm{~mm} \times 108 \mathrm{~mm}$ A new range of high
quality nyoving coil gnatity noving instriments ideal for school experj-
ments and other ments and other
bench applications. $3^{\prime \prime}$ minror scale. The meter move interual working. Availeble in the following ranges:

 | $100 \mu \mathrm{~A}$ | $\ldots$. | $25 \cdot 10$ | 50 V. d.c. | $\cdots$ | 24.85 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 mA | $\ldots .$. | 54.85 | 300 V. d.c. | . | $\mathbf{5 4 . 8 5}$ | $\begin{array}{ll}50-0-50 \mu \mathrm{~A} & 25 \cdot 10 \\ 1-0-1 \mathrm{InA} & 24.85\end{array}$ Dual range $\begin{array}{lllll}1 \mathrm{~A} \text { d.e. } & \ldots . & £ 4.85 & 500 \mathrm{~mA} / 5 \mathrm{~A} \text { d.c. } 25 \cdot 10 \\ 5 \mathrm{~A} \text { d.c. } & \cdots . & 84.85 & 5 \mathrm{~V} / 50 \mathrm{~V} \text { d.c. } & 26.10\end{array}$ 10 V d.e. $\cdot$... 84.85



RP214 REGULATED POWER SUPPLY Solid state. Variable output 0-24V DC up


Brand new. 8 gets of
changeover contacts at changeover contacts at 5 amp rating. 50 p each.
P. \& P. $10 \mathrm{p}(100$ loth 240 ) Quantities a vailable.

AC. or 240 VAC. 200 YA tating. £11.97, carr. 50 p .
MCA. 220 AUTOMATIC VOLTAGE STABILISER
STABILISER
Input $88 \quad 125$ VAC

BH. 001 HEAD SET AND BOOM MICROPHONE
Moving Coll. Ideal for language teaching, commpp. 16 ohns. Miern. phone imp. 200 ohms. £4.82. P. \& P. 15 p .

8-260 General Purpose Bench Mounting


| 1 Amp | $£ 7.00$ |
| :--- | ---: |
| $\mathbf{2 . 5} \mathrm{Amp}$ | $\$ 8.05$ |
| 5 Amp | $£ 11.75$ |
| 8 Amp | $£ 15.90$ |
| 10 Amp | $\mathbf{8 2 6 . 5 0}$ |
| 12 Amp | $£ 28.60$ |
| 20 Amp | $£ 49.00$ |
| 25 Amp | $£ 58.00$ |
| 40 Amp | $£ 82.50$ |



S-260B Panel Mounting
$\begin{array}{lr}1 \mathrm{Amp} & \mathbf{2 7 . 0 0} \\ \mathbf{2 . 5} \mathrm{Amp} & 28.05\end{array}$
Please add postag
INPET 230 V 50/60 CYCLES OUTPUT VARIABLE 0.260 VOLTS Special discounts for quantity

## MULTIMETERS for GUERY purpose，



TS60 POCKET
MULTIMETER MULTIMETER
Hanh－precisioth at luw－cost． 1，000V（ 10,000 op ）．A．C． $15 \mathrm{~V}, 150 \mathrm{~V}, 100 \mathrm{~V}$（ A．C00 Resistance 100 k ohms \＄185．Post 15 p ．

MODEL 1092
TESTMETER
6，000 0．P．V．
$0 / 3 / 15 / 150 / 300 / 1200$ D．C．0／6／30／300／600 $0 / 10 \mathrm{~K} / 1 \mathrm{me}$ g $\Omega$ Decibels -10 to e2．75 each +15 p P．\＆ P


HIOKI MODEL 720X 20,000 O．P．V
$5 / 25 / 100 / 500 / 1000$ $5 / 25 / 100 / 500 / 1000 \mathrm{~V}$
$10 / 50 / 250 / 1000 \mathrm{VAC}$ $50 \mu \mathrm{~A} / 250 \mathrm{ma}$ ． $20 \mathrm{~K} / 2 \mathrm{meg}$


HIOKI MODEL 730X 30,000 O．P．V．Overloar protection．6／30／60／300／ 600／1200 VDC．12／60／120／ $600 / 1200$ VAC． $60 / 2 \mathrm{AA} / 30$ mal $300 \mathrm{~mA} .2 \mathrm{~K} / 200 \mathrm{~K} / 2$ megohm．\＆P． 150


ROUND SCALE TYPE PENCIL TESTER MODEL T．S． 68 Completely portable，simple to nuse pocket sized tester． Ranges $0 / 3 / 30 / 300 \mathrm{~V} A C$
and DC at 2,000 $\begin{array}{lll}\text { and DC at } & 2,000 & \text { o．p．v．} \\ \text { Resistance } \\ 0.20 \mathrm{~K} & \text { olmms．}\end{array}$



LT601 New style 20.000
o．p．v．pocket multi－ o．p．v．pocket multi－
meter． $5 / 25 / 50 / 2501$ neter． $5 / 25 / 50 / 250 /$ $10 / 50 / 100 / 500 / 1000 \mathrm{~L}$ 4．C． $50 \mu \mathrm{~A} / 250 \mathrm{~mA}$ ． $6 \mathrm{~K} / \mathrm{T}$ Heg ohmas．-20 to +22 dB ．
f3．75．Poxt 20 p ．


MODEL TH－12
20.000 o．p．v．Overload pro－ tection．Slide switoh selector
$0 / .25 / 25 / 10 / 50 / 250$
 1000 V ．A．C． $0 / 50 \mu \mathrm{~A} / 25 /$ 250 mA, D．C． $0 / 3 \mathrm{~K} / 30 \mathrm{~K} /$
$300 \mathrm{~K} / 3 \mathrm{meg}-20$ to +50 dB $300 \mathrm{~K} / 3 \mathrm{meg}-20$
$\$ 4.97$. Post 15 p ．

MODEL TE－300
M0．000 O．P．V．Mirror scal overload protection 0／－6／3／15／60／ $1,200 \mathrm{~V}$ A．C． $0 / 30 \mu \mathrm{~A} / 6 \mathrm{~mA} /$
$60 \mathrm{~mA} / 300 \mathrm{~mA} / 600 \mathrm{~mA}$ ． $0 / 8 \mathrm{~K}$ $6 \mathrm{mad} / 300 \mathrm{~mA} / 600 \mathrm{~mA}$ ． $0 / 8 \mathrm{~K} /$
$80 \mathrm{~K} / 800 \mathrm{~K} / 8 \mathrm{meg}$ ohm 80k j800к／8 meg．ohm－20
$+63 \mathrm{db} . \pm 6.97$ ．Н．\＆P．15p．


## MODEL PL 436



MODEL TE－I2
20，000 O．P．V． $0 / 1 \cdot 6 / 6 / 30 / 1: 20$ 600／1，200／3，000／6，000 v．D．C． $0 / 6 / 307120 / 600 / 1,200$ $0 / 604 \mathrm{~A} / 6 / 60 / 600 \mathrm{~mA}$ ． $0 / 6 \mathrm{~K} /$ $600 \mathrm{~K} / 6 \mathrm{Meg} / 60 \mathrm{Meg} . \Omega 50 \mathrm{pF}$
0.2 mFd 0.2 mFd ． 25.97 ．P．\＆ $\mathrm{P}, 17 \mathrm{p}$ ．

MODEL TE－200
20，000 O．P．V．Mirror secale． overload protection． $0 / 5 / 25 / 125 /$ 1，000V．A．C． $0 / 50 \mu \mathrm{~A} / 250 \mathrm{ma}$ ，$/ 60 \mathrm{~K} / 6 \mathrm{meg} \Omega$ ．-20 to +62 d b



## MODEL 500 30,000 O．P．V

30,000 O．P．V．with orer－ Ioad protection mirror scale $0 / 5 / 2 \cdot 5 / 10 / 25 / 100 / 250 / 600 /$
$1,000 v .0 \mathrm{c}$ $\begin{array}{lll}1,000 v . ~ D . C . ~ & 0 / 2.5 / 10 / 25 / \\ 100 / 250 / 500 / 1,000 \mathrm{~V} . & \text { A．C．}\end{array}$ $100 / 250 / 500 / 1,000 \mathrm{~V}$
$0 / 50 \mu \mathrm{~A} / 5 / 50 / 500 \mathrm{~mA}$ amp．D．C．0／60／K／6 Meg．／ $80 \mathrm{Meg} \Omega$
8.87 ．Post paid．

## HIOKI MODEL 750X

50，000 o．D．V． 43 rangen $0-0.9$ A．C． $0-300 \mathrm{~V}$ D．C． $0-3$ to $1,200 \mathrm{y}$ A．C． $0-30 \mu \mathrm{~A} / 300 \mathrm{~mA}$ ． $0-3 \mathrm{~K} / 311$
meg ohms．-10 to $+17 /{ }^{2} \mathrm{l}$ meg． 8 ．Post 20 p ．


HTIOOB4 MULTI－METER

## 100,000 o．p．eurent ranges． <br> Overioad protection． <br> $0 / 5 / 2 \cdot 5 / 10 / 50 / 250 / 500 / 1000 \mathrm{~s}$ <br> 0／2－5／10／50／250／1000 v．A．C． $0 / 10 / 250 / L \mathrm{~A} / 2 \cdot 5 / 25 / 250 \mathrm{MA}$ $10 \mathrm{Map} . \mathrm{D}$ 10 Amp A．C． $0 / 20 \mathrm{~K} / 200 \mathrm{~K} /$ $2 \mathrm{MEG} / 20 \mathrm{MEG}$, e12．50．P．\＆P．25p． <br> 370 WTR MULTI－METER <br> Features A．C．current ranges 20,000 o．p．7． $0 / 5 / 2-5$ $250 / 500$ 1000 V．D．C． <br> $20 / 2.5 / 10 / 50 / 250 / 500 / 1000 \mathrm{~V}$ AC $0 / 50 \mathrm{uA} / 1 / 10 / 100 \mathrm{mal} / 10 \mathrm{~A} A$ <br> $0 / 100 \mathrm{~mA} / 1 / 10 \mathrm{Amp} \mathrm{AC}$ <br> $0 / 5 \mathrm{~K} / 50 \mathrm{~K} / 500 \mathrm{~K} / 5 \mathrm{MEC/}$ <br> － <br> $-20+62 d \mathrm{~b}$. $\mathbf{2 1 5} . \mathrm{P}$ \＆P． 25 p

RUSSIAN 22 RANGE MULTIMETER Model U437 10,000 irst class versatile strument manufactured int U．S．S．R．to the highest gtandards．Ranges：2．5／101 $50 / 250 / 500 / 1000 \mathrm{v}$ D．C． 2.51 10／50／250／500／1000v DC Current $100 \mathrm{wA} / 1 / 10$ $100 \mathrm{~mA} / 1 \mathrm{~A}$ ．Resistance 300 ohms $/ 3 / 30 / 300 \mathrm{~K} / 3 \mathrm{~m} \Omega$ ． Complete with batteries，
test leads，instructions and test leads，instructions and
 sturdy steel carrying case．
OUR PRICE 45.97. P．$P$ ． $2 \overline{0}$,
$20 \mathrm{k} \Omega /$ Volt D．C． $8 \mathrm{k} \Omega /$ Volt AC．Mirror scale． ${ }^{-6 / 3 / 12 / 30 / 120 / 600}$ D．C． $3 / 300120 / 600 / \mathrm{V}$ mA． $10 / 100 \mathrm{~K} / 1 \mathrm{Meg} / 10$ $\mathrm{meg} \Omega-20$ to +4 ed d. \＆6．97．P．\＆P．12p．

TMK MODEL TW－50K 46 ranges，mirror scale 50K／Vol．D．C．5K／Volt A．C．
 $2.6,5,10,25,50,125,225$ ， 500,1000 V，A．C．Volts： $1 \cdot 5$,

$3,5,10,25,50,125,250,500$, $3,5,10,25,50,125,250,500$ ， | $1000 V$ |  |  |
| :--- | :--- | :--- |
| $50 \mu \mathrm{~A}$, | D． $5, ~$ | 5, | $50 \mu \mathrm{~A}, ~$

$500 \mathrm{~mA}, 5,10 \mathrm{amp}, 25, \quad 50,250$,
Resistance： $10 \mathrm{~K}, 100 \mathrm{~K}, 1 \mathrm{MEG}, 10 \mathrm{MEG}$ $\Omega$ ．Dectbels：-20 to $+81 \cdot 5 \mathrm{db}$ \＆8．50．P．\＆P． 17 p ．
MODEL
K22BA
suapension．
Overload
protection
Polarity
Polarity
reversing

$0 / \cdot 5 / 2 \cdot 5 / 15 / 50 / 280$
$500 / 1000 / 2500 \mathrm{~V}$ ．D．E． $0 / 15 / 50 / 150$ $500 / 1000 \mathrm{~V}$ ．A．C． 0 ／50んA／5／50／150／ 50 mind $/ 5 A$ D．C． $0 / 3 K / 300 \mathrm{~K} / 3$ meg．
$£ 8.95$ ．Post $: 20 \mathrm{p}$. 88．95．Post 20 p ．
HIOKO MODEL 700X 100,000 O．P．V．Overload protection．Mirror seale．
$-3 / 6 / 1 \cdot 2 / 1 \cdot 5 / 3 / 6 / 2 / 2 / 30 / 60 /$ $120 / 300 / 6001200 \mathrm{~V}$ DC 1．5／3／5／12／30／60／150／800／600／ 1900 V ．A．C．
10／BOLLA／3／6／30／60／150／300m Mer AMP．DC． $2 \mathrm{M} / 200 \mathrm{~K} / 2$
 ＋6．1B． $\mathbf{5 1 3} 50$ ．P．\＆P． 20 p ．

## MODEL C－7080 EN  5000 V ．D．C． $0 / 2.5 / 10$／ $500 / 250 / 1000 / 5000 \mathrm{~V}$ $\mathrm{~A} C \cdot 0 / 50 / 1 / 1 / 10 / 100$  $-20 \mathrm{tos}+50 \mathrm{~dB}$. E 13.95 ．Pust 35 N. <br> 

## U4312 MULTIMETER

Extremely sturdy instrument for general elecrical use． $667 / 1-5 / 7 \cdot 6 / 30 / 60 / 150 / 300 /$ $600 / 900$ VEC and 75 mV ． $0 /-3 / 1 \cdot 5 / 7 \cdot 5 / 30 / 60 / 150 / 300$ 6001900 VAC．
$0 / 300 \mu A / 1 \cdot 5 / 6 / 15 / 60 / 150$ 600MA／L－0／6 A MP．D．C． $0 / 1 \cdot 6 / 6 / 15 / 60 / 150 / 600 \mathrm{MA}$ 1．5／6 AMP．AC． $0 / 200 \Omega / 3 \mathrm{~K} / 30 \mathrm{~K} \Omega$ Accuraey DC 1\％AC I－5\％


Knife edge polnter．mirror scale．Complete Winstructions es． 50 plus $P$ of $\mathbf{P}$ ．25，leads and

## Selected TEST EOUIPMENT

FTC－40I TRANSISTOR TESTER
Fule capabilities for measuring Equally adaptable for check ing diodes．Supplied complete withis instruchons，battery ant leads．
$\mathbf{~} 7.50$ ．

50．Post 20 p ．

Model S－IOOTR MULTIMETER TRANSISTOR TESTER
100，000 o．p．v．mirror scale／
overload
protection．
$0 / .12 /$ overioad protection．
$-6 / 12 / 30 / 120 / 600 \quad$ DC 0／6／30／120／600．V
$600 \mu \mathrm{~A} / \mathrm{l} 12 / 300 \mathrm{~mA} / 12$ AMP DG．0／10 K／1 MEG／100MEG -20 to $+50 \mathrm{db} .0 \cdot 01-2 \mathrm{MFD}$ ． Alphana，beta and Ico．Complete with batterles，instructions and leads． $813 \cdot 50$ ．P／P 25 p ．


MODEL 449A IN CIRCUIT TRAN－ SISTOR TESTER Checks truè A．C． beta in／out．Checks in $;$ out．Checiss
SCR，etc．Beta
0\％ $2-50$ ．Icbo $0-5000 \mu \mathrm{H} 220 / 240^{-} \mathrm{V} 500$ peration
£17．50．Post $\mathbf{2 5 F}_{5}$ ．

## TE－20D RF SIGNAL GENERATOR



Accurate wide range sig．
nal generator covering
$120 \mathrm{Kc} / \mathrm{s} \quad 500 \mathrm{Mc} / \mathrm{s}$ on 6 bands．Directly cali brated Variable R．F．at－ tenuator，audio output Xtal socket for calibra
tion．
$220 / 240 \mathrm{~V}$ ．A．C Brand new with instruc tions 215 ．Carr． $37 \pm \mathrm{P}$ tions 215. Cart．
Size $140 \times 215 \times 170 \mathrm{~mm}$

MODEL L－55 FET V．0．M．
Input impedance 10 meg ohms． $0 /-3 / 1 \cdot 2 / 6 / 30 /$ $12 / 60 / 120 / 600 \mathrm{~V}$ ．A．C． $0 / 120 \mu A / 120 \mathrm{~mA}$ D．C．
$0 / 1 \mathrm{~K} / 100 \mathrm{~K} / 10 \mathrm{meg}$ f100 meg ohms

 Sengitivity at $100 \mathrm{KHz}, \nabla$ RMS／mam．3－25： Preset triggered sweep 1－3，000usec；free rumning $20.200,000 \mathrm{~Hz}$ in nine ranges Oslibrator pips． $220 \times 360 \times 430 \mathrm{~mm}$ ． $115-230 \mathrm{~V}$ ．AC operation． $\mathbf{8 3 9} \mathbf{0 0}$ ．Carr．paid．

## TO． 3 PORTABLE OSCILLOSCOPE

 3in．tube，Y amp．Sensitivity$0 \cdot 1 \mathrm{p}$ pp／CM．Bandwidth $1 \cdot 5 \mathrm{cps}-1 \cdot 5 \mathrm{MHz}$ ．Input imp． $\begin{array}{llll}2 \mathrm{meg} \\ 2 \mathrm{l} & 25 \mathrm{pF} & \mathrm{X} \text { amp．} \\ \text { gensitivity } & 0.9 \mathrm{v}, & \mathrm{p} \cdot \mathrm{p} / \mathrm{CM} .\end{array}$ $\begin{array}{ll}\text { gensitivity } & 0.9 \mathrm{v}, \quad \mathrm{p} \cdot \mathrm{p} / \mathrm{CM} . \\ \text { Bandwidth } & 1.5 \mathrm{eps}-800 \mathrm{kHz}\end{array}$ Bandwidth $1.5 \mathrm{eps}-800 \mathrm{kHz}$ ．
Input imp． $2 \cdot \mathrm{meg} \Omega 20 \mathrm{pF}$ Input imp． $2 \cdot$ meg $\Omega 20 \mathrm{pF}$ ．
Time bage． 5 ranges 10 cps 300 kEz．Synchronization． Internal／external．Illuminated seale $140 \times 215 \times 380 \mathrm{~mm}$ ．Weight 15 tlb ． 220／240V．A．C．Supplied brand new with handbook． $\mathbf{\$ 4 0 \cdot 0 0 \text { ．Carr．50p．}}$

RUSSIAN CI－16 DOUBLE BEAM OSCILLOSCOPE
5 me／s Pass Band．Separate Y1 and Y2 amplifiers． Rectangular bin．$\times$ 4lin．
C．R．T．Calibrated trig－ gered sweep from 2 u／sec． to 100 milli－sec．per cm ． Free running time base $50 \mathrm{c} / \mathrm{s}-1 \mathrm{mc} / \mathrm{s}$ ．Built in time base calibrator and amplitude calibrator．Supplied complete with all accessories and
f87．Carr．Pald．

TE－16A TRANSISTORISED SIGNAL
 GENERATOR
Ø ranges $400 \mathrm{KHz}-30 \mathrm{mHz}$ An inexpensive instru ment for the handyman Wide easy to read acale． 800 kHz modulation $5 \frac{7}{8} \times 55_{8}^{2} \times 3{ }_{8}^{2} \mathrm{in}$ ．
tions and leads．87．97．Post 25 p
TRANSISTORISED L．C．R．A．C．
MEASURING BRIDGE


BELCO AF－5A SOLID STATE SINE SQUARE WAVE C．R．OSCILLATOR Sine $18 \times 200.000 \mathrm{~Hz}$ ：Square $18 \times 50,000 \mathrm{~Hz}$
 Output max $(10 \mathrm{~KB}$ ohms） Operation in． ternal batteries Athractive
tone case $7 \boldsymbol{2}$
－
$\times$ tone case 7 In $^{\prime \prime} \times$
$5^{n} \times 2^{\prime \prime}$

Price | Price |
| :---: |
| Carr． 17170.50 | MODEL MG－100 SINE SQUARE WAVE AUDIO GENERATOR Range 19－220，000Hz Sine Wave 19－100，000

Hz Square Output Sine or Square Size $180 \times 90 \times 10 \mathrm{P} .9$ to $\mathbf{P}$ ． 417．50．Post 37p


## MODEL AT201

 DECADE ATTENUATOR Frequency range 0－200 KHz ．Attene 200 KHz ．Attenuator
$0-111 \mathrm{db}, 0-1 \mathrm{db}$ $0-111 \mathrm{db}, 0 \cdot 1 \mathrm{db}$ step．
Impedance 600 ohms． Impedance 600 ohms．
Max．input power 30dbm．Size $180 \times 90 \times 5 \mathrm{mmm}$ ． 212．50．Post 37 p ．

TE－65 VALVE VOLTMETER 28 ranges．D．C．volts
$1 \cdot 5-1,500 \mathrm{c}$.
$1.5-1,500 \mathrm{v}$ I． $5-1,500 \mathrm{v}$ ．Resistance up
to 1,000 megohms． $200 /$ to 1,000 megohms． 2007
240 v ．A．O．operation Complete with probe and instructions． \＆17．50．P．\＆P． 30 p.
Additional probes avail－ Additional probes aval－
able：R．F． $22.12 \frac{1}{5}$ ；H．V．
£2．50．
MODEL U4311 SUB－STANDARD MULTI


RANGE VOLT AMMETER
sensitivity 330 ohms／Volt
AC and D．Accuracy
$\mathbf{5 \%}$ D．C． $1 \%$ AC．Beale
I．5／3／7．6／15／30／75
$150 / 300 / 750 \mathrm{~mA} / 1.5 / 3$
$7.5 A M P$ DCO／3／7．5／15
7．5AMP DCO／3／7．5／15
$30 / 75 / 150 / 300$
$750 \mathrm{~mA} / 1.5 / 3 / 7.5$
$\mathrm{AMP} \mathrm{AC} 0 / 75 / 150$
$300 / 750 \mathrm{~m}$
$300 / 750 \mathrm{mV} / 1.5 . / 3$
$/ 150 / 300 / 750 \mathrm{DC}$ $0 / 750 \mathrm{mV} / 1.5 / 3 / 7.5 / 15 / 30 / 75 / 150 / 300$ 1750 V AC．Autoruatic cut out．Supplied com－ plete with test leads，manual and test certif cates．$£ 48 \cdot 00$ ．Post 50 p．

UNR-30

## RECEIVER

4 Bands covering $550 \mathrm{Ke} / \mathrm{s}-30 \mathrm{Mc} / \mathrm{s}$ Speaker $220 / 240$. A.C. Brand new with instructions. \&15.75. Carr. 37 p .


UR-IA SOLID STATE COMMUNICATION RECEIVER
4 Bands covering $550 \mathrm{Kc} / \mathrm{s}-30 \mathrm{Mc} / \mathrm{s}$. FHT S Meter. Variable BFO for SSB, Built-in Speaker, Bandspread, Sensitivity Control. $220 / 240$ v. A.C. or 12 v . D.C. $124^{\prime \prime} \times 47^{\prime \prime} \times 7^{\prime \prime}$ Brand new with instructions. $£ 25$. Carr. 37 p .

SKYWOOD CX203 COMMUNICATION RECEIVER


Solid state, Coverage on 5 bands 200-420 KHz and -55 to 30 MHz . Illuminated slide rule dial. Bandspread. Aerial tuning BFO, grated speaker and phone socket. Operation $220 / 240 \mathrm{v}$ AC or 12 v DC. Size $325 \times 266 \times 150$ £3250. Carr. 50 p .

LAFAYETTE HA-600 SOLID STATE Receiver
 General coverage
$150-400 \mathrm{Kc} / \mathrm{s}$ $\begin{array}{lr}150-400 \mathrm{Kc} / \mathrm{s} & 550 \\ \mathrm{Kc} / \mathrm{s}-30 \mathrm{Mc} / \mathrm{s} . & \text { FET }\end{array}$ Ke/s-30Mc/s. FET
front end. 2 mech. fitters,
detector,
vroduct B.F.O., nolse limiter. S. Meter, Bandspread, RF Gain. $15^{\prime \prime} \times$
$99^{\prime \prime} \times 8 \frac{1}{2}^{\prime \prime} .18$ ibs. $220 / 240 \mathrm{v}$. A.C. Or 12 v. D. Bra.d new with instructions. 850, Carr. 50p


TRIO 9R59DS COMMUNI-
CATION RECEIVER

4 band covering $550 \mathrm{Ke} / \mathrm{s}$. to
$30 \mathrm{Mc} / \mathrm{s}$. con. 30 Me/s. con-
tinuous and electrical bandspread on $10,15,20,40$ and
80 metres. 8 valve plus 7 diode circuit. $4 / 8$ ohm output and phone jack. SSB-CW.
ANL. Variable BFO. S meter. sep. band. Apread dial. IF frequency $455 \mathrm{Kc} / \mathrm{s}$. bandoutput 1.5 w Variable RF and 4 F gain controls $115 / 250 v$. A.C. Size: $7^{*} \times 13^{\prime \prime} \times 10^{*}$ with instruction manual. $£ 49 \cdot 50$. Carr. Paid.


EMI LOUDSPEAKERS
Model 350. $13^{\prime \prime} \times 8^{\prime \prime}$ with single tweeter/crossover. ${ }^{20-}$
$20,000 \mathrm{~Hz}$. 15 watt RMS. Avallable 8 or 15 ohms. $£ 7 \cdot 25$
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or 15 ohms. $\& 8 \cdot 62$ P. \& P. 25 p .

HONEYWELL
DIGITAL
OLTMETER
Cian be panel or bench mounted.
 sures 1 volt D.C.
but can be used to measure a wide range of optional plug in cards. Specifications with racy: $\pm 0 \cdot 2, \pm 1$ digit. Resolution: ImF. Number of digits: 3 plas fourth overrange digit. Overrange: $100 \%$ (up to $1-999$ ). Input impedance: 1000 Meg ohm, Measuring cycle: 1 per second. Adjustment: Automatic zeroing, full scale adjustment against an internal roference voltage. Overload: to 100 v . D.C. Input: Fully floating ( 3 poles). Input power. $110-230 \mathrm{v}$. A.C. $50 / 60$ cycles. Overall size: $5 \frac{5}{3}$ in. $\times 2$ 2 $13 / 16 i n . \times 83 / 16 \mathrm{inl}$. AFA1LABLE TEED. $\mathbf{\$ 3 5}$-50. Carr. 50p.

## SINcLaIR IC-12 List Price

 £2.98
## 1) OURTPRICE $£ 1 \cdot 80$

P. \& P. 10 p

## SINCLAIR EQUIPMENT <br> Project 60 Package <br> 

 offers.$2 \times \mathrm{Z}$
power amplifier, stereo 60 pre-an!. PZó power supply. \&15.95 Carr. 37p. Or with PZ6 pnwer supply 618 -00 Carr. $37 \mathrm{p} .2 \times 250$ supply. 220'25. Carr. 37 p .
Transformer for PZ8. e2.97 extra
Add to any of the above $£ 4.45 \mathrm{fm}$ active filter unit and 11800 for pair of Q16 speakers. All other Sinclair products in stock.
2000 Anp 221.95 Carr, 372 p .: 3000 Amp £28-50 Carr. $379 . ;$ Neoteric Ainp $£ 43.95$. Carr. 37p ICl 2 21-80 p. \& p. 10p.
NEW PROJ FCT $605-\mathbf{8 2 0 . 9 7 . ~ V a r t . ~}$

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GOODMANS AXIOM 301 Hi Fi $12^{\sim} 20$ watt twin cou Hull range speaker. $30-16,000$
$\mathrm{~Hz}, \quad 16,500$ gauss. 8 imper 16,500 gauss. 8 ohms
imper boxed.
(List price £21.72) OUR


SPECIAL OFFER:
Matchell pair of stereo bookshelf speakers. Deluxe teak reneered
finish. Size $14 t^{-} \times 9^{-} \times$ finish. Size $14 \frac{1}{2}$ 7f". $^{\text {RMS. }}{ }^{8} \quad$ ohns. ${ }^{8}{ }^{8}$ watt watt Complete with TIN lead. £12.95 pr, Carr. 50p.


HA-10 STEREO HEADPHONE AMPLIFIER
All silicon transator amplitier oper-
ates from magnetic, ceramic or tuner inputs with twin stereo headphone outputs and separate volume battery Inputs $5 \mathrm{MU} / 100 \mathrm{MU}$ Output 50 MF結-97. P. \& P. 15 p .

SPECIAL PURCHASE: NEAT G30J STATIC BALANCE PICK-UP ARMS


Identical specification to NEAT G30 arm but with rwo-tone chrome and black finish. Complete with head shell, pick up rest and plug in phono leads.
BRAND NEW-
FULLY GUARANTEED
ONLY £8.95. P. \& P. 25p.
ARF-300 AF/RF Signal Generator


All transistorised, compact, fully port.
able. AF sine wave is Hz to $2 \geq 0 \mathrm{KHz}$ 18 Hz to 100 KHz Output sine / square 10 v . P-P. RF 100 KHz to 200 MHz . Output 15 . maximum. Opera tions and leads. $828 \cdot 95$. Post 50 p .

1021 STEREO LISTENING STATION Fain balancing and oud selection ot oudspeakers with additional facility forstereo headphone
switching. 2


DOLBY SYSTEM NOISE

lmproves the perforiance of cassetfe and semin-professional recorders. Reluces tape
hiss by 3 dB at $600 \mathrm{~Hz}, 6 \mathrm{dH}$ at 1200 Hz and 10 dB for all trequencies above 3000 Hz Controls for input levels and noise reduction on record and replay. 2 meters for Dolby response; 20 Hz to $15 \mathrm{kHz} \pm 1 \mathrm{~dB} \quad 19 \mathrm{kHz}$ $-35 d B$. Size $155^{\prime \prime} \times 9^{\prime \prime} \times 33^{\prime \prime}$. AC $200 / 250 \mathrm{~V}$ switching. 2 gain
slide switeh, stereo controls, Rpeaker miof slide switch,
headphone sockets. $6^{\prime \prime} \times 4^{\prime \prime} \times 22^{\prime \prime} .22 \cdot 25$, P. \&P. 15 p .

MP7 MIXER PREAMPLIFIER

uts each with
mbiridual gain
coutrols enabling
$1.91^{\prime \prime} \times 5^{\prime \prime} \times 3^{\prime \prime}$
facilities. Battery operater:
Inputs Mies : $3 \times 3 \mathrm{mV} 50 \mathrm{~K}: ~$
$\times 3 \mathrm{mV}$
600 ohm .
Phono meg. 4 mV 50 K . Phono ceramie 100 mF l meg. Output 250 mV 100 K . £8.87. P. \& P. 20p.


TE-1035 STEREO HEADPHONES Low cost high performance stereo headphones.
Foann rubber ear eups Foam rubbe rear eups 8 oim impedance 25 $18,000 \mathrm{~Hz}$. With lead and stereo jack plug. ONL
$£ 1.97$. P. \& P, l2p.
NEW GARRARD MODULES


Popular range of Garrard decks with Shure cartridge fitted in deluxe plimth with hinged lid.

##  AP76 Module/Mio-6 AP96 Module/M75-6 <br> Carr. 50 p extr <br> 233.80

Carr. 50 extra any item. $\mathbf{2 5 2 . 6 0}$

## HOSIDEN DH-08S DE-LUXE STEREO

## HEADPHONES

eatures wigue meh nical 2 way units and atter alljustable leven chitrols. 8 ohn impeiamue $20-20,000 c \mathrm{cs}$. lad $\&$ ete with spring 97-97. P. \& P. 12p.

## AKAIEARGAINS <br> SUPER MONEY SAVING OFFERSBUY NOW WHILE STOCKS LAST: BRAND NEW AND FULLY GUARANTEED 1721 Tape Rec. $\mathbf{X} 5000$ The X5000 Tape Rec. GX370 Tape Deck 4000DS Tape Deck 4000DS Dust Cover ${ }^{4000} \mathbf{2 0 0}$ Tape Deck X201D Tape Deck X221D Tape Deck GX280D Tape Deck X1810D Tape/8 track Deck GX1900D Tape/Cass Deck X2000sD Tape/Cas/8 Tr CR81 8 track Rec. <br> CR8ID 8 track Rec. <br> CR80DS 8 track/Rec 8 -track <br> CRROSS 8 track syatem <br>  <br> (arriage 50p extra. (Recorders \& Decks 75p)

 Wonderful value formance conbinedAdjustable headband. 8 ohm impedance. . 20-12,000 eps. Complete with lead and plug.
ONLY $£ 2.37$. $P$. P. 12p. 2.37 . P. \&


SPECIAL OFFER! ROTEL RH700 STEREO HEADPHONES 20-20,000Hz. 8-16 ohm List 69.95 ). OUR PR. PRICE
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TRANSISTORISED FM TUNER
 HIGH QUASTITY TUNER, QUALIT ONLY $6 \times 4 \times 2 \frac{1}{2}$ in.
$3 \quad$ I.F. $\quad$ stages. Double tuned dis criminator, Ample output to feed most amplifters. Operates on 9 battery. Coverage tastic value for moner. 86.371 P \& P 1210 Stereo multiplex adaptors $£ 4 \cdot 97 \frac{1}{2}$
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Sensitive, soft ear Magnetic, impedance ohms. £1-97, P. \& P. 15p

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WALKIE TALKIES
Industrial quality in robust metal casesBaltery operation. Volume and squelch button. Telescopic aerial. Complete carrying cases.
2 channel 2 channel
300 mW $\quad 52.50 \quad \begin{aligned} & \text { Pair. } \\ & \text { Post } 50 \mathrm{p}\end{aligned}$ 3 channel 879.50 Pair.
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Also see previous page and opposite page

## FANTASTIC OFFER!


$17+17$ watts rms stereo amplifier with inputs for Magnetic and List price 859.50 Crystal phono, Tuner, Tape, Aux and Tape Monitor. Outputs for two pairs of stereo speakers and Tape. Stereo headphone socket. Full range of controls including loudness control, serateh filter, ete. Size $13^{\prime \prime} \times 9 \underline{\underline{t}}^{\prime \prime} \times 3 \underline{\underline{a}}^{\prime \prime}$. Unrepeatable offer-limited £39.95 atocks!


LEAK DELTA 30 SYSTEM


TELETON SAQ206B SYSTEM

$\underset{\text { PRICE }}{ } £ 54.40$
Carr.
fl. 50
Amplifier only, £22.95. Carr. 50p.


Tone, volume and balance controls. Track selector. Complete with matched pair of stereo speakers, connections and fittings. ONLY £15.95. Post 30p. (Thus is example only).

## BH. 001 HEAD <br> SET AND BOOM <br> MICROPHONE

Moving Coil. Ideal for language teaching, communications. Headphone imp. 15 ohms. Micro-
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##  Linton Amplifier, Linton Turntable, pair of Linton 2 speakers and all leads. <br> OUR $£ 106.00$ <br> Carr. and PRICE EIO6.00 Ins, $f 1 \cdot 25$ LINTON RECEIVER SYSTEM $£ 155 \cdot 00$. Carr. \& Ins. £1-25. <br> AMSTRAD 8000 II SYSTEM

Amstrad $8000117+7$
Watt a mplifier. BSR
MP60, plinth and cover, Goldring G800 cartridge, pair of Apollo all leads.
${ }_{\text {PRIC }} £ 48.25$
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B.S.R. TDBS 8:TRACK STEREO TAPE PLAYER DECK
Integrated preamps (output 125 mV ) to feed into any stereo amplifier. Automatic and manual programme selector. 4 pole synchronous motor. $210 / 240 \mathrm{~V}$.
OUR PRICE $£ 16.25$

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3414 STEREO
TUNER
AMPLIFIER
TURNTABL:
UNIT
$10+10$
Five push buttons with separate scales for pre-tuning to desired FM station. Housed in a handsome walnut finished cabinet with BSR PI28/MP60 record deck with Goldring G800H stereo magnetic cartridge. Offered complete with cover and a pair of matching Medway Speakers, size $18^{\prime \prime} \times 11^{\prime \prime} \times 8^{\prime \prime}$. TODAY'S VALUE AT LEAST $£ 125$ !


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SAVE UP TO 33 $\frac{1}{3} \%$ OR MORE
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As used in world famous system. $5^{\prime \prime}$ dia. Impedance $4 / 8$ ohms. High flux cms. Brand new $\& 1-50$. Carr. 37p.

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Tdeal for home office, stores, factories, etc. Supplied complete with bat-
teries
cable and teries, cable and
iree instructions.
2 Station, $\mathbf{2 2 \cdot 8 7} 3$ Station $85 \cdot 25$, P. \& P. 15p
4 Station \&6-62, P. \& P. 17p
CREDIT TERMS FOR CALLERS ACCESS \& BARCLAYCARD WELCOME

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All mail orders received on or after April Ist. will be subject to Value Added Tax.
THIS MUST BE ADDED TO THE TOTAL VALUE OF GOODS ORDERED (INCLUDING POSTAGE/CARRIAGE) AND ENCLOSED WITH ORDER.

RECORD DECKS (P.\& f. 50p)
B.S
C11
C12
C13
610
810
210
MP
MP
MP
MP
610
510
HT
HT
HT
HT
810


BD1 Kit BD1 Chosis $\begin{gathered}£ 10.90 \\ \mathbf{8 1 3 . 6 0}\end{gathered}$ BD1/SAU2/Plinth BD2/SA U2/Chassis BD2/SAU2/Plinth/C GARRARD $\mathbf{5 3 3 - 8 5}$ $2025 \mathrm{~T} / \mathrm{C}$ Stereo $£ 8.50$ SP25 III/Acos ONALD
£4.97
$\mathbf{8 6} .50$
 P60/G800 88.95
$£ 12.95$

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| :--- | :--- |
| 185 |  |

 ZERO 100A $\begin{array}{ll}\text { C25.05 } \\ \text { G28.05 }\end{array}$ ZERO 1009 ZERO 100S E86.95 | ZERO 100S MOd/ |
| :--- |
| M93-E |
| $52 \cdot 60$ | M93-E

M75-6 Mode/
M38-75 GOODMAN TD100 Teak $\mathbf{5 5 1 - 9 5}$ TD100 White 858.25
GOLDR



| GL69/2 | $\mathbf{8 1 8 . 5 0}$ |
| :--- | :--- |
| GL72 | $\mathbf{2 2 0 . 9 5}$ |

$\begin{array}{ll}\text { GLIT2 } & £ 20.95 \\ \text { GLIT2/P } & £ 27.50\end{array}$
$\begin{array}{lr}\text { QLi/2/P } & \text { 227.60 } \\ \text { Plinth } 69 / 72 & \mathbf{E 7} \cdot 62 \\ \text { LID } 72 & \mathbf{8 3 . 2 5}\end{array}$

$$
\begin{aligned}
& \text { GLD } 72 \\
& \text { GLD }
\end{aligned}
$$

$$
\begin{aligned}
& \text { GL75 } \\
& \text { GL75P } \\
& \text { Plinth } 75
\end{aligned}
$$

$$
\begin{aligned}
& \text { GLidor } \\
& \text { Plinth } 75 \\
& \text { LID } 75
\end{aligned}
$$

$$
\begin{aligned}
& \text { LID } 75 \\
& \text { G99 }
\end{aligned}
$$

$$
\begin{array}{ll}
\text { G99 } & \mathbf{5 1 9 . 2 5} \\
\text { GL85P/C } & \mathbf{5 5 6 . 9 5} \\
\text { G101P/C } & 520.50
\end{array}
$$

$$
\begin{aligned}
& \text { G99 } \\
& \text { GL85 }
\end{aligned}
$$

 LEAK

## MICRO-SEIKI

 MR111 829.50 $\underset{\text { Cover }}{\text { MR111 }}$ Plinth \&
## PHILIPS

$\begin{array}{ll}\text { GA105 } & \text { \&18.95 } \\ \text { GA160 } & £ 27.00\end{array}$ $\begin{array}{ll}\text { GA160 } & \begin{array}{ll}\mathbf{~} 27 \cdot 00 \\ \text { GA308 Teak } & 824 \cdot 50\end{array}\end{array}$ GA308PU Teak GA212 $\quad \mathbf{5 5 6}$.75
PIONEER
$\begin{array}{ll}\text { PL12D } & \mathbf{8 3 4}-50 \\ \text { PL15C } & \mathbf{8 5 1} .36\end{array}$

$\begin{array}{ll}\text { PL41D } \\ \text { PL61 } & \begin{array}{l}\mathbf{\$ 1 1 8 . 5 0} \\ \mathbf{£ 1 1 9 . 9 5}\end{array}\end{array}$
THORENS
TD125 II $686 \cdot 50$
TD125AB II 299.95 $\begin{array}{lr}\text { TX25 } & \mathbf{8 6 . 9 5} \\ \text { TD } 60 \mathrm{C} & \mathbf{8 5 6 . 9 5}\end{array}$
 WHARFEDALE
Linton Iurntable
$\mathbf{5 2 8 . 9 5}$

| PLINTHS \& COVERS (P, \& P. 50p) |  |
| :---: | :---: |
| Budget SPR5, etc. | 83-20 |
| Budget AP76/Zero 1008 | 24.50 |
| Budget B.S.R. | £3.25 |
| SME 2000 System | £34.50 |

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Decks supplied with
Decks suppled cartridge
ready wired in plinth ready wired in plinth with cover.

| GA |  |  |  |
| :---: | :---: | :---: | :---: |
| 2025 TC/9TAHCD |  |  | 212-75 |
| SP25 IIT/G800 |  |  | 818-50 |
| SP25 1II/M75-6 |  |  | ¢18.50 |
| SP25 IIT/M44-7 | . |  | 219.75 |
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| AP76/G800 | . | . | £27.95 |
| AP76/M75-6 | $\cdots$ |  | ¢30.25 |
| AP76/M55E |  |  | £30.50 |
| A P76/M75EJ |  |  | 232.50 |
| AP76/G800E |  |  | 980.75 |
| AP76/M44E |  |  | 230.50 |
| AP76/M75ED |  |  | £38.95 |
| B.S.E. McDONALD |  |  |  |
| MP60/G800 | $\cdots$ | . | ${ }^{217} 50$ |
| MP60/M44-7 |  | . | ¢18.95 |
| MP60/M44-E | . | . | 219.95 |
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| GL751G800 |  |  | 288.50 |
| G1.75/G800E | . | . | £42.50 |



The Space Age WORLD RECORD award winning $7^{\prime} 6^{\prime \prime}$ long All Band JOYSTICK VFA Antenna - JOYMATCH A.T.U. TRIO RECEIVER or TRANSCEIVER - Matching HEADPHONES - A COMPLETE STATION IN ONE COMPACT PACKAGE - THE FLAT DWELLER'S DREAM!

1972 JOYSTICK VFA £11-40(A); JOYMATCH ATU 111 tor 111A £11-40 (B); JOYMATCH ATU LO-Z500 £16.45 (C).

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(All including postage, packing, accessories \& insurance.)

PARTRIDGE PACKAGE NO. 1


Complete Gen. Coverage
Station
C60.50
Station ... .. .. $\mathbf{6 6 0 - 5 0}$

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JR599 recalver, (D) abouv .. £185.09
Complete Amateur RX Station $£ \mathbf{£ 8 5 \cdot 0 0}$

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## Brilliant new technical specifications

Input impedance $100 \mathrm{~K} \Omega$
Input (for 30 w into $8 \Omega$ ) 400 mV
Signal to noise ratio, referred to full o/p at $30 \mathrm{v} \mathrm{HT} \mathrm{80dB} \mathrm{or} \mathrm{better}$
Distortion $0.02 \%$ up to 20 W at $8 \Omega$. See curve Frequency response 10 Hz to more than $200 \mathrm{KHz} \pm 1 \mathrm{~dB}$
Max. supply voltage 45 v ( $4 \Omega$ to $8 \Omega$ speakers)
( $50 \vee 15 \Omega$ speakers only)
Min. supply voltage 9 v
Load impedance-minimum : $4 \Omega$ at 45 v HT Load impedance - maximum : safe on open circuit


The $Z .30$ provides excellent facilities for the constructor requiring a high fidelity audio system of less power than that available from Z.50's Using a power supply of 35 volts. $Z .30$ will deliver 15 watts RMS into 8 ohms. or 20 watts RMS into 3 ohms using 30 volts. Total harmonic distortion is a fantastically low $0.02 \%$ at 15 watts into 8 ohms with signal to noise ratio better than 70 dB unweighted. Input sensitivity 250 mV into 100 K ohms. Size $80 \times 57 \times 13 \mathrm{~mm}\left(3 \frac{1}{8} \times 2 \frac{1}{4} \times \frac{1}{2}\right) Z .30 . Z .50$ and Z.50 MK. 2 modules are compatible and interchangeable.

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## Typical Project 60 applications

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| Simple battery record player | Z.30 | Crystal P.U.. 12 V battery volume control, etc. | £4.48 |
| Mains powered record player | Z.30, PZ. 5 | Crystal or ceramic P.U. volume control, etc. | £9.45 |
| 12W. RMS continuous sine wave stereo amp. for average needs | $\begin{aligned} & 2 \times Z .30 \mathrm{~s}, \text { Stereo } \\ & 60 ; \text { PZ.5 } \end{aligned}$ | Crystal, ceramic or mag. <br> P.U., F.M. Tuner, etc. | £23.90 |
| 25W. RMS continuous sine wave stereo amp. using low efficiency (high ,performance) speakers | $\begin{aligned} & 2 \times \text { Z.30s, Stereo } \\ & 60 ; \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc. | £26.90 |
| 80 W . ( 3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms) | $2 \times Z .50$ s, Stereo 60; PZ.8, mains transformer | As above | £34.88 |
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SPECIFICATIONS-Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz . Sensitivity: $7 \mu \vee$ for lock-in over full deviation. Squelch level: Typically $20 \mu \vee$. Signal to noise ratio: $>65 \mathrm{~dB}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}$ ( 1 dB ). Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation Stereo decoder operating level: $2 \mu \mathrm{~V}$. Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S. maximum Operating voltage: $25-30 \mathrm{VDC}$. Indicators: Sterec on; tuning. $\$ i z e: ~ 93 \times 40 \times 207 \mathrm{~mm}$.

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