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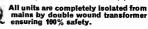
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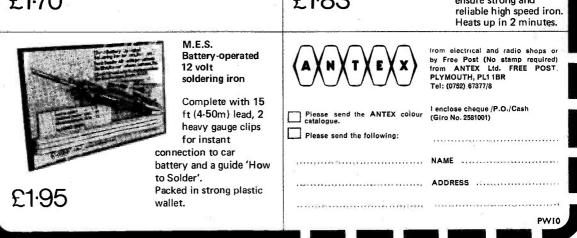


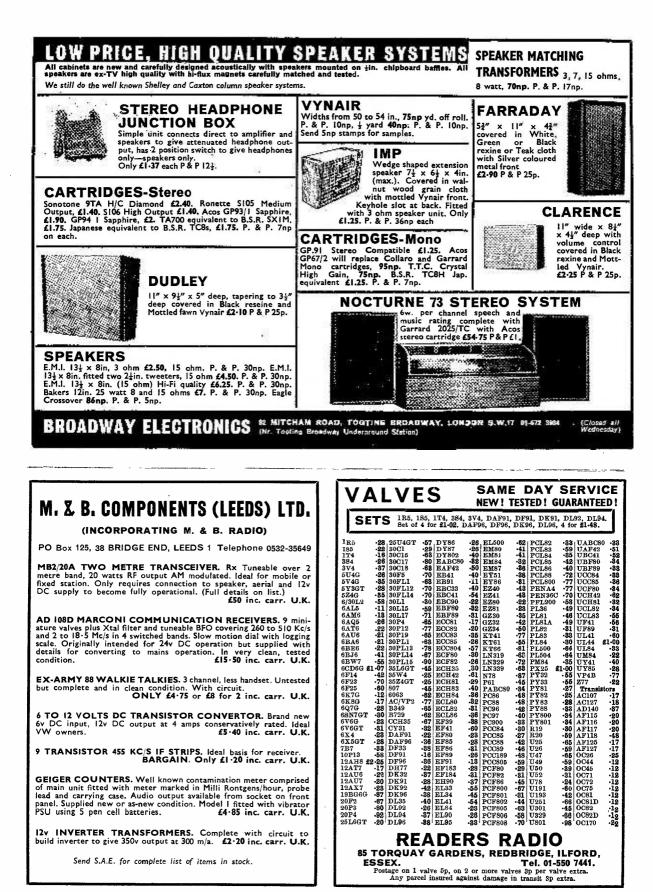
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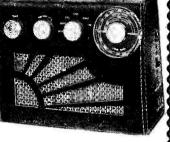
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6 Tunable Wave-bands: MW, LW, SW1, SW2, Trawier bend piles an extra . M.W. band for easief tuning of Luxembourg, etc. Sensitive forrite rod aerial and telescopic aerial for Sant 2 layes. Sin. Speaker. 8 stages-6 transistors, atc. Attractive luxime Micro-Alloy R.F. Thansistors, based knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with red grille, dial and black knobs with poliabes with red grille, dial and black knobs with red grille, dial and black

Total building costs £3-98 P. P. & (Overseas P. & P. £1)



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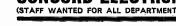
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There are two stereo amplifiers-the R100 for ceramic cartridges, the RIOI for magnetic. Both incorporate FETs (FIELD EFFECT TRANSISTORS), just like top-priced units. FETs give you more of the signal you want, and almost none of the background hiss you don't. Both units have a jack socket to plug in headphones and there's a separate output for tape recorder. Filters (an unusual feature in this price range) and tone controls give a wide range of bass and treble adjustment which compensate for input deficiencies and domestic acoustic conditions.

PRICES SYSTEM 1

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Viscount III R101 amplifier 2 x Duo Type II speakers,	£22.00+90p p&p £14.00+£2 p&p
Garrard SP25 Mk. III with M cart ridge plinth and cover	£23.00+£1.50
Total	£59.00
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SYSTEM 2 Viscount R101 amplifier 2 x Duo Type III speakers	£22.00+90p p&p £32.00+ £3 p&p
Garrard SP25 Mk. III with M cartridge, plinth and cover	£23·00+£1·50 p&p
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SYSTEM 3

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SPEAKERS Duo Type II

Size $17" \times 10\frac{3}{4}" \times 6\frac{3}{4}"$. Drive unit $13" \times 8"$ with parasitic tweeter. Max. power 10 watts. 3 ohms. Simulated Teak cabinet. £14 pair+£2 p&p. Duo Type III Size $23\frac{1}{2}'' \times 11\frac{1}{2}'' \times 9\frac{1}{2}''$. Drive unit $13\frac{1}{2}'' \times 8\frac{1}{4}''$ with H.F. speaker. Max. power 20 watts at 3 ohms. Freq. range 20Hz to 20kHz. Teak veneer cabinet. £32 pair+£3 p&p.

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SPECIFICATION R100/101

14 watts per channel into 3 to 4 ohms. Total distortion @ 10W @ 1kHz 0.1%. P.U.1 (for ceramic cartridges) 150mV into 3 Meg. P.U.2 (for magnetic cartridges) 4mV @ 1kHz into 47K. equalised within $\pm 1dB$ R.I.A.A. Radio 150mV into 220K. (Sensitivities given at full power). Tape out facilities; headphone socket, power out 250mW per channel. Tone controls and filter characteristics. Bass: +12dB to -17dB @ 60Hz. Bass filter: 6dB per octave cut. Treble control: treble +12dB to -12dB @ 15kHz. Treble filter: 12dB per octave. Signal to noise ratio: (all controls at max) R101-P.U.1 and radio-65dB. P.U.2. -58dB. RIOO same as RIOI but P.U.2 (for crystal cartridges) 450mV into 3 Meg. Cross talk better than -35dB on all inputs. Overload characteristics better than 26dB on all inputs. Size $13\frac{1}{2}" \times 9" \times 3\frac{1}{2}"$.



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to set up and can be relied upon to come over with all the quality and power you need. Output Power: 45 watts R.M.S. (Sine wave drive). Frequency response: -3dB points 30Hz at 18KHz. Total distortion: less than 2% at rated output. Signal to noise ratio: better than 60dB. Speaker Impedance: 3. 8 or 15 ohms. Bass Control Range: \pm 13dB at 60Hz. Treble Control Range: \pm 12dB at 10 KHz. Inputs: 4 inputs at 5mV into 470K. Each pair of inputs controlled by separate volume control. 2 inputs at 200mV into 470K. To protect the output valves, the incorporated fail safe circuit will enable the amplifier to be used at half power. SPEAKERS! Size 20" \times 20" \times 10" incorporating 12" heavy duty 25 watt high flux, quality loudspeaker with cast frame. Cabinets attractively finished in two tone colour scheme—Black and grey.

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U28	Experimenters' assortment of integrated circuits, untested.		Q38 7 PNP trans. 4x2N3703, 3x2N3702 0
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U29	10 1-Amp SCR's TO-5 can up to 600 PIV CR81/25-600		Q41 3 Plastic NPN TO-18 2N3904 0
U31	20 Sil. Planar NPN trans. low noise amp 2N3707		Q42 6 NPN trans. 2N5172 0 Q43 7 BC107 NPN trans 0
U32	25 Zener diodes 400mW D07 case mixed volts, 3-18	0.50	Q44 7 NPN trans. 4×BC108, 3×BC109 0
U33	15 Plastic case 1 amp silicon rectifiers IN4000 series	0-50	Q45 3 BC113 NPN TO-18 trans 0 Q46 3 BC115 NPN TO-5 trans 0
U34	30 Sil. PNP alloy trans. TO-5 BCY26, 28302/4		Q47 6 NPN high gain 3 x BC167, 3 x BC168 0
U35	25 Sil. planar trans. PNP TO-18 2N2906		Q48 4 BCY70 NFN trans. TO-18 0 Q49 4 NPN trans. 2×BFY51, 2×BFY52 0
U36	25 Sil. planar NPN trans. TO-5 BFY50/51/52	0-50	Q50 7 BSY28 NPN switch TO-18 0
U37	30 Sil. alloy trans. SO-2 PNP, OC200 28322		Q51 7 BSY95A NPN trans. 300MH2 0 Q52 8 BY100 type sil. rect. 1
U38	20 Fast switching sil. trans. NPN, 400Mc/s 2N3011		Q53 25 Sil. & germ. trans. mixed all
U39		0-50	marked new 1-
U40	10 Dual trans. 6 lead TO-5 2N2060	0.50	ADINTED CIRCUITS EV CONSULT
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BP00 = 7400 Quad 2-input NAND gate 0.15 0.14 0.12 BP01 = 7401 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BP02 = 7402 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BP03 = 7402 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BP04 = 7404 Twistopen collector output) 0.15 0.14 0.12 BP05 = 7406 Her Impere collector output) 0.15 0.14 0.12 BP10 = 7410 Triple 3-input pos. NAND gates 0.15 0.14 0.12 BP13 = 7413 Dual 4-input Schmitt trigger 0.29 0.26 0.24 BP43 = 7446 Dual 4-input pos. NAND gates 0.15 0.14 0.12 BP44 = 7447 DC to decimal nixe driver 0.47 0.46 0.88 BP44 = 7448 BCD-to-seven-segment decoderghriver 0.47 0.48 0.48 BP45 = 7431 Dual 2-winde 2-input and-or-invert gates 0.15 0.14 0.12 BP45 = 7431 Dual 2-winde 2-input and-or-invert gates 0.15 <td>BI-PAK</td> <td>Alexilian Barran to a Description</td> <td>Price</td> <td>and giy.</td> <td>prices</td>	BI-PAK	Alexilian Barran to a Description	Price	and giy.	prices
BF01 = 7401 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BF02 = 7402 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BF03 = 7402 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BF04 = 7404 Her Inverter (with open-collector output) 0.15 0.14 0.12 BF05 = 7406 Her Inverter (with open-collector output) 0.15 0.14 0.12 BF10 = 7410 Triple 3-input pos. NAND gates 0.15 0.14 0.12 BF13 = 7413 Dual 4-input Schmitt trigger 0.29 0.26 0.24 BF14 = 7440 Dual 4-input Schmitt trigger 0.27 0.66 0.58 BF44 = 7441 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF44 = 7446 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF45 = 7448 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF46 = 7446 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF46 = 7446 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF46	order No.	Similar Types to:-Description	1-24	25-99	100 up
BF01 = 7401 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BF02 = 7402 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BF03 = 7402 Quad 2-input pos. NAND gate 0.15 0.14 0.12 BF04 = 7404 Her Inverter (with open-collector output) 0.15 0.14 0.12 BF05 = 7406 Her Inverter (with open-collector output) 0.15 0.14 0.12 BF10 = 7410 Triple 3-input pos. NAND gates 0.15 0.14 0.12 BF13 = 7413 Dual 4-input Schmitt trigger 0.29 0.26 0.24 BF14 = 7440 Dual 4-input Schmitt trigger 0.27 0.66 0.58 BF44 = 7441 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF44 = 7446 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF45 = 7448 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF46 = 7446 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF46 = 7446 BCD-to-seven-segment decoder/driver 0.67 0.64 0.58 BF46	BP00 = 7400	Quad 2-input NAND gate	0.15		5D 0.19
BF02 7402 Quad 2-input pos. NADE gates 0.15 0.14 0.12 BF04 7404 Hoz Invertes 0.015 0.14 0.12 BF04 7404 Hoz Invertes 0.015 0.14 0.12 BF04 7404 Hoz Invertes 0.015 0.14 0.12 BF10 7410 Dual 4-input schmitt higgets 0.15 0.14 0.12 BF20 7400 Dual 4-input schmitt higgets 0.15 0.14 0.12 BF20 7400 Dual 4-input schmitt higgets 0.15 0.14 0.12 BF40 7441 Dual 4-input schmitt higgets 0.15 0.14 0.12 BF41 7441 DCD to decimal accoder furver 0.67 0.64 0.58 BF42 7442 BCD to decimal decoder furver 0.67 0.64 0.58 BF43 7445 BCD-to-seven-segment decoder furver 0.15 0.14 0.12 BF44 7444 BCD-to-seven-segment decoder furver 0.16 0.14	BP01 = 7401	Quad 2-input nos. NAND gate		0.7.8	0.45
Br04 Cirkit Appen collector output) 0.15 0.14 0.12 Br05 7.406 Her Inverters 0.15 0.14 0.12 Br10 7.410 Triple 3-input pos. NAND gates 0.15 0.14 0.12 Br13 Triple 3-input pos. NAND gates 0.15 0.14 0.12 Br20 7420 Dual 4-input Schmitt trigger 0.28 0.22 0.22 0.22 0.22 Br20 7420 Dual 4-input Schmitt trigger 0.28 0.22 0.23 0.24 0.15 0.14 0.13 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15	-	(with open collector output)		0.14	0.12
Br04 Cirkit Appen collector output) 0.15 0.14 0.12 Br05 7.406 Her Inverters 0.15 0.14 0.12 Br10 7.410 Triple 3-input pos. NAND gates 0.15 0.14 0.12 Br13 Triple 3-input pos. NAND gates 0.15 0.14 0.12 Br20 7420 Dual 4-input Schmitt trigger 0.28 0.22 0.22 0.22 0.22 Br20 7420 Dual 4-input Schmitt trigger 0.28 0.22 0.23 0.24 0.15 0.14 0.13 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15 0.14 0.12 0.15	BP02 = 7402 BP09 = 7402	Quad 2 input pos. NOR gates	0.15	0.14	0.12
BP06 = 7404 Her Inverters 0.15 0.14 0.12 BP10 = 7410 Triple 3-input pos. NAND gates 0.15 0.14 0.12 BP13 = 7413 Dual 4-input pos. NAND gates 0.15 0.14 0.12 BP13 = 7413 Dual 4-input pos. NAND gates 0.15 0.14 0.12 BP30 = 7430 S-input pos. NAND gates 0.15 0.14 0.12 BP41 = 7440 Dual 4-input pos. NAND gates 0.15 0.14 0.12 BP44 = 7440 BCD to decimal inkie diriver 0.67 0.64 0.68 BP44 = 7446 BCD-to-seven-segment decoder, diriver 0.67 0.64 0.68 BP47 = 7447 BCD-to-seven-segment decoder, diriver 0.15 0.14 0.12 BP50 = 7450 Expandable dual 2-input and-or-invert 0.15 0.14 0.12 BP53 = 7453 Quad 2-input and-or-invert gates 0.15 0.14 0.12 BP70 = 7470 Master alave J-K flip-flop 0.63 0.62 0.62 BP71 = 7472 Master alave J-K flip-flop 0.63 0.62 0.62 BP72 = 7472 Master alave J-K flip-flop </td <td>DI 03 4 1403</td> <td>(with open collector output)</td> <td>0.15</td> <td>0.14</td> <td>. 10</td>	DI 03 4 1403	(with open collector output)	0.15	0.14	. 10
BP05 = 7405 Her. Inverter (with open-collector 0.15 0.14 0.12 BP10 = 7410 Triple 3-input pos. NAND gates 0.15 0.14 0.12 BP13 = 7410 Dual 4-input Schmitt trigger 0.28 0.28 0.28 0.28 BP20 = 7420 Dual 4-input Schmitt trigger 0.28 0.28 0.28 0.28 BP40 = 7440 Dual 4-input schmitt trigger 0.28 0.28 0.28 0.28 BP44 = 7442 Dual 4-input schmitt trigger 0.28 0.28 0.28 0.28 BP44 = 7442 BCD to decimal theoder(4-in) 0.67 0.64 0.88 BP45 = 7445 BCD-to-seven-segment decoder/drivers 0.97 0.94 0.88 BP50 = 7450 Expandable dual 2-input and-or-invert gates 0.15 0.14 0.12 BP54 = 7454 Suide 2-input expandable and-or-invert gates 0.15 0.14 0.12 0.14 0.12 BP76 = 7450 Dual 4-input expandable and-or-invert gates 0.15 0.14 0.12 BP76 = 7454 Jual Master slave J-K flip-flop 0.37 0.45 0.42 BP77 = 7472 <td>BP04 = 7404</td> <td>Her Inverters</td> <td>0.15</td> <td></td> <td></td>	BP04 = 7404	Her Inverters	0.15		
BP10 = 7410 output) cols NAND gates 0-15 0-14 0-12 BP13 = 7413 Dual 4-input pos. NAND gates 0-28	BP05 = 7405		0.10	0.14	0.12
BP10 = 7410 Triple 3-input pos. NAND gates 0-15 0-14 0-12 BP13 = 7420 Dual 4-input pos. NAND gates 0-15 0-14 0-12 BP20 = 7420 Dual 4-input pos. NAND pates 0-15 0-14 0-12 BP40 = 7440 Dual 4-input pos. NAND pates 0-15 0-14 0-12 BP40 = 7440 Dual 4-input pos. NAND buffers 0-67 0-64 0-58 BP44 = 7442 BCD to decimal decoder (4-10 lines, 0-67 0-64 0-58 BP44 = 7447 BCD-to-sven-aggment decoder/drivers 0-07 0-94 0-88 BP45 = 7448 BCD-to-sven-aggment decoder/drivers 0-15 0-14 0-12 BP50 = 7450 Expandable dual 2-input and-or-invert gates 0-15 0-14 0-12 BP54 = 7454 Sugle phase 1-K flip-flop 0-29 0-28 0-28 0-29 BP70 = 7450 Dual 4-input synander 0-15 0-14 0-12 BP74 = 7454 4 wide 2-input and-or-invert gates 0-15 0-14 0-12 BP76 = 7450 Dual 4-input synander 0-15 0-14 0-12 BP77 = 7472 Bua		output)	0.15	0.14	0.12
B121a -1413 Duki 4-input pos. NAND gates 0-28		Triple 3-input pos. NAND gates	0.15	0.14	0.12
BP30 = 7430 S-input pos. NAND pates 0.15 0.14 0.12 BP40 = 7440 BQD to decimal nixie driver 0.67 0.64 0.55 BP42 = 7440 BQD to decimal nixie driver 0.67 0.64 0.55 BP44 = 7441 BQD to decimal nixie driver 0.67 0.64 0.55 BP45 = 7445 BQD-to-seven-segment decoder/driver 0.97 0.94 0.88 BP46 = 7446 BQD-to-seven-segment decoder/driver 0.97 0.94 0.88 BP50 = 7450 Expandable dull 2-input and-or-invert 0.97 0.94 0.88 BP51 = 7451 Dual 2-wide 2-input and-or-invert gates 0.15 0.14 0.12 BP53 = 7453 Quad 4-input expander 0.15 0.14 0.12 BP54 = 7454 4-wide 2-input and-or-invert gates 0.15 0.14 0.12 BP76 = 7460 Dual 4-input expander 0.15 0.14 0.12 BP77 = 7472 Master alave J-K flip-flop 0.59 0.24 BP78 = 7445 Dual J-K with pre-set and clear 0.47 0.45 BP74 = 7474 Dual J-K with pre-set and clear 0.47	BP13 = 7413	Dual 4-input Schmitt trigger	0.29		0.24
BF44 = 7441 BCD to decimal decoder (d-10 lines, 1 of 10) 0.67 0.64 0.55 BF46 = 7446 BCD-to-seven-segment decoder/driver 0.07 0.94 0.88 BF47 = 7447 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BF48 = 7448 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BF48 = 7448 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BF50 = 7450 Expandable dull 2-input and-or-invert 0.15 0.14 0.12 BF51 = 7451 Dual 2-wide 2-input and-or-invert gates 0.15 0.14 0.12 BF60 = 7460 Dual 4-input expander 0.15 0.14 0.12 BF72 = 7470 Misgle phase J-K flip-flop 0.82 0.98 0.24 BF73 = 7473 Dual J-K with pre-set and clear 0.47 0.45 0.42 BF74 = 7474 Dual J-K with pre-set and clear 0.47 0.48 0.48 BF88 = 7488 Quad 111 adder 0.97 0.94 0.88 BF88 = 7448 Quad 111 adder 0.97 0.94 0.88 BF88 = 7448 Quad	BP20 = 7420 BP20 = 7420	LUZI 4-Input pos. NAND gates			0.12
BF44 = 7441 BCD to decimal decoder (d-10 lines, 1 of 10) 0.67 0.64 0.55 BF46 = 7446 BCD-to-seven-segment decoder/driver 0.07 0.94 0.88 BF47 = 7447 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BF48 = 7448 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BF48 = 7448 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BF50 = 7450 Expandable dull 2-input and-or-invert 0.15 0.14 0.12 BF51 = 7451 Dual 2-wide 2-input and-or-invert gates 0.15 0.14 0.12 BF60 = 7460 Dual 4-input expander 0.15 0.14 0.12 BF72 = 7470 Misgle phase J-K flip-flop 0.82 0.98 0.24 BF73 = 7473 Dual J-K with pre-set and clear 0.47 0.45 0.42 BF74 = 7474 Dual J-K with pre-set and clear 0.47 0.48 0.48 BF88 = 7488 Quad 111 adder 0.97 0.94 0.88 BF88 = 7448 Quad 111 adder 0.97 0.94 0.88 BF88 = 7448 Quad	BP40 = 7440	Dual 4-input nos NAND buffere		0-14	0.12
BP42 = 742 BCD to decimal decoder (4-10 lines, 0-67 0-64 0-58 Br46 = 7446 BCD-to-seven-segment decoder/driver 2-00 1-75 1-50 Br47 = 7447 BCD-to-seven-segment decoder/driver 0-97 0-94 0-88 Br50 = 7450 Expandable dual 2-input and-or-invert 0-97 0-94 0-88 Br53 = 7453 Quad 2-input expandable and-or-invert 0-15 0-14 0-12 Br54 = 7454 Wiret 0-15 0-14 0-12 Br54 = 7453 Quad 2-input expander 0-15 0-14 0-12 Br57 = 7473 Mage raise d-K flip-flop 0-28 0-28 0-24 Br77 = 7473 Dual Master slave J-K flip-flop 0-37 0-35 0-32 Br76 = 7475 Dual J-K with pre-set and clear 0-47 0-46 0-88 Br89 = 7481 Gatori funder 0-47 0-46 0-88 Br89 = 7481 Gatori funder 0-47 0-46 0-88 Br99 = 7490 Gatori funder 0-67 0-44 0-88	BP41 = 7441	BCD to decimal nixie driver			
BP46 7446 BCD-to-sever-segment decoder/driver 0.01 7.5 1.50 BP48 7447 BCD-to-sever-segment decoder/driver 0.97 0.94 0.88 BP48 7448 BCD-to-sever-segment decoder/driver 0.97 0.94 0.88 BP50 7430 Expandable dual 2-input and-or-invert 0.15 0.14 0.12 BP51 7431 Dual 4-input expandable and-or-invert 0.15 0.14 0.12 BP54 7454 4-wide 2-input expander 0.15 0.14 0.12 BP50 7460 Dual 4-input expander 0.15 0.14 0.12 BP70 7470 Single-phase J-K filp-flop 0.89 0.92 0.92 BP74 7471 Dual J Kw filp-flop 0.67 0.63 0.83 B88 B78 5.03 B88 0.92 0.92 0.92 0.92 0.92 0.92 0.93 0.93 0.93 0.93 0.93 0.93 0.93 0.93 0.93 0.93 0.93	BP42 = 7442	BCD to decimal decoder (4-10 lines.	0.01	0.64	0.99
BP445 = 7446 BCD-to-seven-segment decoder/driver 2.00 1.75 1.60 BP47 = 7447 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BP50 = 7450 Expandable dual 2-input and-or-invert 0.97 0.94 0.88 BP51 = 7451 Dual 2-wide 2-input and-or-invert 0.15 0.14 0.12 BP53 = 7453 Quad 2-input expandable and-or-invert 0.15 0.14 0.12 BP54 = 7474 Worts -0.15 0.14 0.12 BP57 = 7470 Single-phase J-K filp-flop 0.23 0.24 0.22 BP70 = 7460 Dual J-kropt expander 0.15 0.14 0.12 BP72 = 7470 Single-phase J-K filp-flop 0.23 0.22 0.24 BP74 = 7474 Dual J Katt filp-flop 0.37 0.35 0.32 BP76 = 7475 Dual J Katt filp-flop 0.43 0.43 0.40 0.38 BP80 = 7480 Gated full adders 0.67 0.64 0.38 0.82 0.82 BP80 = 7480 Gated full adders 0.67 0.64 0.58 0.88 0.88 0.88		1 of 10)	0.87	0-64	0.58
BP48 7445 BCD-to-sever-segment decoderdriver segment decoderdriver Erpandable dual 2-input and-or- invert 0-97 0-94 0-88 BP50 = 7450 Dual 2-wide 2-input and-or- invert 0-15 0-14 0-12 BP51 = 7451 Dual 2-wide 2-input and-or- invert 0-15 0-14 0-12 BP54 = 7453 Quad 2-input expandable and-or- invert 0-15 0-14 0-12 BP64 = 7454 4-wide 2-input and-or-invert gates 0-15 0-14 0-12 BP74 = 7473 Single-phase J-K flip-flop 0-29 0-28 0-24 BP73 = 7473 Dual Master slave J-K flip-flop 0-37 0-35 0-32 BP74 = 7474 Dual J-K with pre-set and clear 0-47 0-40 0-38 BP76 = 7475 Quai latch 0-47 0-44 0-88 BP88 = 7480 Gated full adders 0-47 0-44 0-88 BP89 = 7490 Dad L-R with pre-set and clear 0-47 0-44 0-88 BP89 = 7481 Gottar radius adders 0-47 0-46 0-88 <td< td=""><td></td><td>BCD-to-seven-segment decoder driver</td><td></td><td></td><td></td></td<>		BCD-to-seven-segment decoder driver			
BP48 = 7445 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 BP50 = 7450 Expandable dull 2-input and-or-invert 0.15 0.14 0.12 BP51 = 7451 Dual 2-wide 2-input and-or-invert 0.15 0.14 0.12 BP53 = 7453 Quad 2-input expandable and-or-invert 0.15 0.14 0.12 BP54 = 7454 4-wide 2-input and-or-invert gates 0.15 0.14 0.12 BP64 = 7460 Dual 4-input expander 0.15 0.14 0.12 BP76 = 7470 Single-phase 1-K filp-flop 0.29 0.28 0.24 BP77 = 7472 Master slave 1-K filp-flop 0.37 0.35 0.32 BP76 = 7475 Quad but more stand clear 0.43 0.40 0.38 BP80 = 7480 Gated full adders 0.47 0.46 0.58 BP81 = 7481 16-bit read/write memory 0.97 0.94 0.83 BP82 = 7482 Quad full adders 0.67 0.64 0.58 BP82 = 7486 Quad 2-input exclusive NOR gates 0.27 0.94 0.83 BP84 = 7484 Dual entry 4-bit shift register	BP47 = 7447	BCD-seven-segment decoder/drivers			
BP50 = 7450 Ergandable dual 2-input and-or-invert 0-15 0-14 0-12 BP51 = 7451 Dual 2-wide 2-input and-or-invert 0-15 0-14 0-12 BP53 = 7453 Quad 2-input expandable and-or-invert gates 0-15 0-14 0-12 BP54 = 7454 4-wide 2-input and-or-invert gates 0-15 0-14 0-12 BP66 = 7460 Dual 4-input expander 0-15 0-14 0-12 BP70 = 7470 Single-phase J-K flip-flop 0-29 0-28 0-24 BP71 = 7472 Dual Master slave J-K flip-flop 0-37 0-35 0-32 BP76 = 7472 Quait latch 0-47 0-43 0-40 0-33 BP76 = 7474 Dual J-K with pre-set and clear 0-47 0-44 0-83 BP78 = 7481 Gated full adders 0-97 0-94 0-88 BP80 = 7480 Got fait rediving more 0-67 0-64 0-58 BP91 = 7491 Got fait fait register 0-67 0-64 0-58 BP80 = 7480 Got fait fait register 0-67 0-64 0-58 BP91 = 7491 Got fait fait register	BB49 - 7448	(15 v outputs)			
invertinvert0.150.140.12BP51 = 7451Dual 2-wide 2-input and-or-invert0.150.140.12BP53 = 7453Quad 2-input expandable and-or-invert0.150.140.12BP64 = 74544-wide 2-input and-or-invertigates0.150.140.12BP60 = 7460Dual 4-input expander0.150.140.12BP70 = 7470Single-phase J-K filp-flop0.290.280.28BP71 = 7472Master alave J-K filp-flop0.370.360.32BP74 = 7474Dual Master alave J-K filp-flop0.370.360.32BP75 = 7475Qual Master alave J-K filp-flop0.370.450.42BP80 = 7480Gatef full adders0.470.460.43BP80 = 7480Gatef full adders0.470.460.43BP88 = 7483Quad full adders0.670.640.63BP88 = 7484Quad full adders0.670.640.58BP89 = 7490BCD decade counter0.670.640.58BP99 = 7492Divide-by-twelve counters0.670.640.58BP99 = 7493BCD decade counter0.670.640.58BP90 = 7490BCD decade counter0.670.640.58BP91 = 74101Shift up-down ahift register0.770.740.68BP93 = 7493Dual entry 4-bit shift register0.770.740.68BP94 = 7494Dual entry 4-bit shift register0.770.740.68BP94 = 74104<	BP50 - 7450	Expendable dual to interest	0.97	0.94	0.88
BF51 = 7451 Dual 2-wide 2-input and-or-invert 0.15 0.14 0.12 Br53 = 7453 Quad 2-input expandable and-or-invert 0.15 0.14 0.12 Br54 = 7454 4-wide 2-input and-or-invert gates 0.15 0.14 0.12 Br74 = 7470 Single-phase J-K flip-flop 0.29 0.28 0.24 Br71 = 7472 Master slave J-K flip-flop 0.29 0.28 0.24 Br73 = 7473 Dual Master slave J-K flip-flop 0.37 0.35 0.32 Br74 = 7474 Dual J-K with pre-set and clear 0.43 0.43 0.43 Br78 = 7474 Dual J-K with pre-set and clear 0.47 0.46 0.58 Br88 = 7480 Gated full adders 0.97 0.94 0.88 Br88 = 7482 Qual full adders 0.97 0.94 0.88 Br88 = 7483 Qual full adders 0.97 0.94 0.88 Br88 = 7484 Qual full adders 0.97 0.94 0.88 Br99 = 7491 Abit register 0.67 0.64 0.58 Br99 = 7492 Pobl decade counter 0.67 0.64 <td< td=""><td>D100 - 1400</td><td>invert</td><td>0.16</td><td>0.14</td><td>A 10</td></td<>	D100 - 1400	invert	0.16	0.14	A 10
gates gates <t< td=""><td>BP51 = 7451</td><td></td><td>0.10</td><td>0.14</td><td>0.12</td></t<>	BP51 = 7451		0.10	0.14	0.12
BF25 = 7453 Quad 2-input expandable and-or- invert 0.15 0.14 0.12 BF260 = 7460 Dual 4-input expander 0.15 0.14 0.12 BF270 = 7470 Single-phase J-K flip-flop 0.29 0.28 0.24 BF71 = 7472 Master slave J-K flip-flop 0.29 0.28 0.24 BF73 = 7472 Muster slave J-K flip-flop 0.37 0.35 0.32 BF76 = 7474 Dual J-K with pre-set and clear 0.43 0.40 0.33 BF76 = 7475 Quait latch 0.47 0.46 0.43 BF76 = 7476 Dual J-K with pre-set and clear 0.43 0.40 0.33 BF88 = 7480 Gated full adders 0.47 0.40 0.33 BF88 = 7482 Quait bit register 0.47 0.44 0.46 BF88 = 7483 Quait J-limpt exclusive NOR gates 0.37 0.40 0.38 BF89 = 7484 Quait bit register 0.47 0.44 0.58 BF910 = 7401 B-liters differegister 0.67 0.64 0.58 BF92 = 7492 Duai entry 4-bit shift register 0.67 0.64			0.15	0-14	0.12
BP54 7454 Invert 0-15 0-14 0-12 BP60 7460 Dual 4-input expander 0-15 0-14 0-12 BP60 7470 Single-phase 1-K flip-flop .0.29 0-28 0-24 BP72 7473 Dual Master slave 1-K flip-flop .0.37 0-35 0-32 BP74 7474 Dual Master slave 1-K flip-flop .0.47 0-46 0-42 BP76 7473 Dual J-K with pre-set and clear 0-43 0-40 0-33 BP80 7480 Gated full adders .0.47 0-44 0-58 BP81 7481 16-bit read/write memory 0-37 0-36 0-32 BP80 7486 Quad 2-input exclusive NOR gates 0-30 0-28 BP80 7486 Quad 1-14 adder 1-10 1-05 0-58 BP80 74496 BCD decade counter .0.67 0-64 0-58 BP90 7491 Bott isplit register .0.77 0	BP53 = 7453	Quad 2-input expandable and-or-	0 10	0.11	0.10
B260 = 7460 4-wide 2-input and-or-invert gates 0-15 0-14 0-12 B270 = 7470 Single-phase J-K flip-flop 0-29 0-28 0-24 B272 = 7470 Master slave J-K flip-flop 0-29 0-28 0-24 B273 = 7472 Muster slave J-K flip-flop 0-37 0-35 0-32 B274 = 7474 Dual Master slave J-K flip-flop 0-37 0-35 0-32 B276 = 7474 Dual J-K with pre-set and clear 0-47 0-46 0-43 B278 = 7481 Gated tull adders 0-47 0-44 0-83 B280 = 7481 Gated tull adders 0-47 0-44 0-88 B280 = 7482 2-bit binary full adders 0-47 0-44 0-88 B280 = 7482 2-bit binary full adders 0-97 0-44 0-88 B280 = 7482 2-bit binary full adders 0-97 0-44 0-88 B280 = 7483 4-bit binary conners 0-67 0-64 0-58 B290 = 7492 Dovide-by-twelve conners 0-67 0-64 0-58 B290 = 7493 4-bit bindt register 0-77 0-74 0		invert	0-15	0.14	0.12
BP70 = 7470 Single-phase J-K flip-flop 0.59 0.52 0.54 BP72 = 7472 Master alave J-K flip-flop 0.59 0.52 0.52 BP73 = 7473 Dual Master alave J-K flip-flop 0.57 0.55 0.52 BP76 = 7474 Dual J-K with pre-set and clear 0.47 0.45 0.42 BP76 = 7475 Qual latch 0.47 0.46 0.43 BP76 = 7476 Dual J-K with pre-set and clear 0.47 0.40 0.53 BP80 = 7480 Gated full adders 0.67 0.40 0.53 BP81 = 7481 16-bit read/write memory 0.67 0.40 0.58 BP82 = 7483 Quad 2-input exclusive NOR gates 0.32 0.30 0.28 BP83 = 7486 Quad 2-input exclusive NOR gates 0.32 0.40 0.58 BP93 = 7490 BCD decade counter 0.67 0.64 0.55 BP13 = 7491 8-bit shilt register 0.77 0.74 0.68 BP32 = 7484 4-bit bit memory in particle with shift register 0.77 0.74 0.68 BP104 = 74100 Shife bitshight heft particle withiff	BP54 = 7454	4-wide 2-input and or-invert gates	0.15	0-14	
BP16 7476 Gain J.K. with pre-set and clear 0.43 0.43 0.42 BP80 7480 Gated full addres 0.67 0.54 0.58 BP80 7481 16-bit read/writh addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP86 7466 Quad full addres 0.67 0.54 0.58 BP90 7490 BCD decade counter 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP95 7494 Dual entry 4-bit shit register 0.77 0.74 0.68 BP96 7496 5-bit bit adile latches 1.75 1.65 1.55 BP104 74104 Single J-K flip flop equivalent 9001 seties 0.67 0.94 0.68 BP105 74104 Single J-K flip flops 0.47 0.74 <td>BP60 = 7460</td> <td>Dual 4-input expander</td> <td></td> <td></td> <td>0.12</td>	BP60 = 7460	Dual 4-input expander			0.12
BP16 7476 Gain J.K. with pre-set and clear 0.43 0.43 0.42 BP80 7480 Gated full addres 0.67 0.54 0.58 BP80 7481 16-bit read/writh addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP86 7466 Quad full addres 0.67 0.54 0.58 BP90 7490 BCD decade counter 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP95 7494 Dual entry 4-bit shit register 0.77 0.74 0.68 BP96 7496 5-bit bit adile latches 1.75 1.65 1.55 BP104 74104 Single J-K flip flop equivalent 9001 seties 0.67 0.94 0.68 BP105 74104 Single J-K flip flops 0.47 0.74 <td>BP70 = 7470 BP70 = 7470</td> <td>Single-phase J-K flip-flop</td> <td>0.29</td> <td>0-26</td> <td>0.24</td>	BP70 = 7470 BP70 = 7470	Single-phase J-K flip-flop	0.29	0-26	0.24
BP16 7476 Gain J.K. with pre-set and clear 0.43 0.43 0.42 BP80 7480 Gated full addres 0.67 0.54 0.58 BP80 7481 16-bit read/writh addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP86 7466 Quad full addres 0.67 0.54 0.58 BP90 7490 BCD decade counter 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP95 7494 Dual entry 4-bit shit register 0.77 0.74 0.68 BP96 7496 5-bit bit adile latches 1.75 1.65 1.55 BP104 74104 Single J-K flip flop equivalent 9001 seties 0.67 0.94 0.68 BP105 74104 Single J-K flip flops 0.47 0.74 <td>BP12 = 7412 BP22 - 2472</td> <td>Master slave J-K flip-flop</td> <td></td> <td></td> <td></td>	BP12 = 7412 BP22 - 2472	Master slave J-K flip-flop			
BP16 7476 Gain J.K. with pre-set and clear 0.43 0.43 0.42 BP80 7480 Gated full addres 0.67 0.54 0.58 BP80 7481 16-bit read/writh addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP81 7482 Quad full addres 0.67 0.54 0.58 BP86 7466 Quad full addres 0.67 0.54 0.58 BP90 7490 BCD decade counter 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP92 7493 4-bit bit register 0.67 0.54 0.58 BP95 7494 Dual entry 4-bit shit register 0.77 0.74 0.68 BP96 7496 5-bit bit adile latches 1.75 1.65 1.55 BP104 74104 Single J-K flip flop equivalent 9001 seties 0.67 0.94 0.68 BP105 74104 Single J-K flip flops 0.47 0.74 <td>BP74 = 7474</td> <td>Dual Master slave J-K nip-nop</td> <td>0-37</td> <td>0-35</td> <td></td>	BP74 = 7474	Dual Master slave J-K nip-nop	0-37	0-35	
P 76 = 7476 Dual J-K with pre-set and clear 0.43 0.44 <td< td=""><td>BP75 = 7475</td><td>Quad latch</td><td>0.87</td><td></td><td>0-82</td></td<>	BP75 = 7475	Quad latch	0.87		0-82
BP30 = 7480 Gated tull adders 0.47 0.64 0.58 BP31 = 7481 16-bit read/write memory 0.87 0.94 0.88 BP32 = 7482 2-bit binary full adders 0.97 0.94 0.88 BP38 = 7485 Quad 2-input exclusive NOR gates 0.32 0.30 0.28 BP38 = 7485 Quad 2-input exclusive NOR gates 0.32 0.30 0.28 BP59 = 7491 BOI decade counter 0.67 0.64 0.55 BP90 = 7493 BOI decade counter 0.67 0.64 0.55 BP90 = 7493 Dola entry 4-bit trift register 0.67 0.64 0.55 BP90 = 7493 Dola entry 4-bit trift register 0.77 0.74 0.68 BP104 = 74104 Shife J-K flip flop equivalent 9001 1.75 1.65 1.55 BP105 = 74105 Shife J-K flip flop equivalent 9001 0.47 0.94 0.88 BP104 = 74104 Shife J-K flip flop s 0.47 0.94 0.88 BP105 = 74105 Shife J-K flip flop s 0.47 0.94 0.88 BP104 = 74104 Shife J-K flip flop s 0.47	BP76 - 7476		0.49	0.40	0.99
$ \begin{array}{llllllllllllllllllllllllllllllllllll$	BP80 = 7480		0.67	0.64	0.58
$ \begin{array}{llllllllllllllllllllllllllllllllllll$	BP81 = 7481	16-bit read/write memory	0.97		
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	BP82 = 7482	2-bit binary full adders	0.97		
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	BP83 = 7483		1.10		0-95
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	$BP86 \simeq 7486$	Quad 2-input exclusive NOR gates	0-32		0.28
$ \begin{array}{c} \hline 3 \mbox{P} 9 = 7492 & 0hride-ty restricts 0.67 & 0.64 & 0.58 \\ \mbox{P} 9 = 7493 & 0hride-ty restricts 0.67 & 0.64 & 0.58 \\ \mbox{P} 9 = 7493 & 0hride-ty restricts 0.67 & 0.64 & 0.58 \\ \mbox{P} 9 = 7494 & Dual entry 4-bit shift register 0.67 & 0.64 & 0.58 \\ \mbox{P} 9 = 7495 & 0hrit up-down shift register 0.77 & 0.74 & 0.88 \\ \mbox{P} 9 = 7496 & 7496 & 5-bit garallel in garallel out shift 0.77 & 0.74 & 0.88 \\ \mbox{P} 9 = 7496 & 7496 & 5-bit bit stable latches 0.77 & 0.74 & 0.88 \\ \mbox{P} 9 = 74100 & 8-bit bitshabe latches 0.77 & 0.74 & 0.88 \\ \mbox{P} 10 = 74100 & 8-bit bitshabe latches 0.77 & 0.74 & 0.88 \\ \mbox{B} P100 = 74100 & 8-bit bitshabe latches 0.77 & 0.94 & 0.88 \\ \mbox{B} P105 = 74105 & 8ingle J-K flip flop equivalent 9001 \\ & series 0.47 & 0.94 & 0.88 & 0.83 \\ \mbox{B} P110 = 74110 & Dual data lock-out flip-flops 0.47 & 0.94 & 0.88 \\ \mbox{B} P110 = 74110 & Hex set-reset latches 0.47 & 0.48 & 0.58 \\ \mbox{B} P111 = 74110 & Hex set-reset latches 0.47 & 0.48 & 0.58 \\ \mbox{B} P112 = 74110 & Hex set-reset latches 0.47 & 0.48 & 0.58 \\ \mbox{B} P141 = 74110 & Hex set-reset latches 0.47 & 0.44 & 0.58 \\ \mbox{B} P141 = 74145 & BC-bt-decimal decoder/driver 0/C & 1.50 & 1.40 & 1.30 \\ \mbox{B} P131 = 74154 & Hex bert of late latches 0.47 & 0.48 & 0.58 \\ \mbox{B} P134 = 74155 & Dual 4-line-to-line data 1.20 & 1.10 & 0.45 \\ \mbox{B} P134 = 74155 & Dual 2-to 4-line decoder / Hiver 0/C & 1.50 & 1.40 & 1.30 \\ \mbox{B} P134 = 74156 & Dual 2-to 4-line decoder 1.40 & 1.50 & 1.60 \\ \mbox{B} P134 = 74166 & Spin. 4-bit bin decoder 0/C & 1.40 & 1.50 & 1.20 \\ \mbox{B} P136 = 74166 & Spin. 4-bit line decoder 0/C & 1.40 & 1.80 & 1.20 \\ \mbox{B} P190 = 74190 & Sync. up-down decade counter & 1.80 & 1.70 & 1.60 \\ \mbox{B} P190 = 74190 & Sync. up-down decade counter & 1.80 & 1.70 & 1.60 \\ \mbox{B} P190 = 74190 & Sync. binary up-down counter (tow clock lines) & & & & & & & & & & & & &$	BP90 = 7490 BP01 = 7490	BCD decade counter	0.67		0-58
$ \begin{array}{llllllllllllllllllllllllllllllllllll$	BP00 = 7400	8-Dit split registers	0.87	0.84	
Dr. 19 F. 190 10.11 parallel in parallel out shift BP100 74100 Stit bistable latches 1.75 1.65 1.55 BP104 74104 Single J-K filp-fop equivalent 9000 0.97 0.94 0.88 BP105 74105 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74105 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74107 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74101 Outat master since filp-fops 0.97 0.94 0.88 0.83 BP110 74110 Outat master since filp-fops 0.95 0.55 0.56 0.50 BP111 74112 Monestable multivibarior 0.67 0.64 0.58 BP141 74121 Monestable multivibarior 0.67 0.64 0.58 BP143 74143 BCD-to-decimal decoder/driver 0.67 0.64 0.58 BP145 74153 Dual 4-line-to-1-line data 1.90 0.95 0.90 BP1543	BP02 - 7402	thit bluery counters			
Dr. 19 F. 190 10.11 parallel in parallel out shift BP100 74100 Stit bistable latches 1.75 1.65 1.55 BP104 74104 Single J-K filp-fop equivalent 9000 0.97 0.94 0.88 BP105 74105 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74105 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74107 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74101 Outat master since filp-fops 0.97 0.94 0.88 0.83 BP110 74110 Outat master since filp-fops 0.95 0.55 0.56 0.50 BP111 74112 Monestable multivibarior 0.67 0.64 0.58 BP141 74121 Monestable multivibarior 0.67 0.64 0.58 BP143 74143 BCD-to-decimal decoder/driver 0.67 0.64 0.58 BP145 74153 Dual 4-line-to-1-line data 1.90 0.95 0.90 BP1543	BP94 = 7494	Dual entry 4-hit shift register	0.77	0.74	
Dr. 19 F. 190 10.11 parallel in parallel out shift BP100 74100 Stit bistable latches 1.75 1.65 1.55 BP104 74104 Single J-K filp-fop equivalent 9000 0.97 0.94 0.88 BP105 74105 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74105 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74107 Single J-K filp-fop equivalent 9001 0.97 0.94 0.88 BP105 74101 Outat master since filp-fops 0.97 0.94 0.88 0.83 BP110 74110 Outat master since filp-fops 0.95 0.55 0.56 0.50 BP111 74112 Monestable multivibarior 0.67 0.64 0.58 BP141 74121 Monestable multivibarior 0.67 0.64 0.58 BP143 74143 BCD-to-decimal decoder/driver 0.67 0.64 0.58 BP145 74153 Dual 4-line-to-1-line data 1.90 0.95 0.90 BP1543	BP95 = 7495	4-bit up-down shift register	0.77	0.74	
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	BP96 = 7496	o-oit parallel in parallel out shift-		0.11	0.00
$ \begin{array}{c} \text{BP104} = 74104 & \text{Single J-K filp-flop equivalent 9000} \\ \text{series} & 0.47 & 0.94 & 0.68 \\ \text{BP105} = 74105 & \text{Single J-K filp flop equivalent 9001} \\ \text{series} & 0.47 & 0.94 & 0.88 \\ \text{BP107} = 74107 & \text{Dual Master slave filp-flops} & 0.47 & 0.94 & 0.88 \\ \text{BP107} = 74107 & \text{Dual Master slave filp-flops} & 0.47 & 0.94 & 0.88 \\ \text{BP107} = 74110 & \text{date master slave filp-flops} & 0.45 & 0.58 & 0.50 \\ \text{BP111} = 74111 & \text{Dual data lock-out filp-flops} & 0.45 & 0.58 & 0.50 \\ \text{BP111} = 74111 & \text{Dual data lock-out filp-flops} & 0.45 & 0.58 & 0.50 \\ \text{BP111} = 74111 & \text{Dual data lock-out filp-flops} & 0.467 & 0.64 & 0.58 \\ \text{BP131} = 74112 & \text{Monestable multivibrators} & 0.47 & 0.48 & 0.58 \\ \text{BP141} = 74112 & \text{Monestable multivibrators} & 0.47 & 0.64 & 0.58 \\ \text{BP141} = 74141 & \text{BCD-to-decimal decoder/(fiver 0/C & 1.50 & 1.40 & 1.30 \\ \text{BP131} = 74151 & 8.016 & \text{data selector} & 1.80 & 1.70 & 1.60 \\ \text{BP135} = 74153 & 10.616 & \text{data selector} & 1.80 & 1.70 & 1.60 \\ \text{BP136} = 74150 & \text{Dual 2- to 4-line decoder} & 1.40 & 1.30 & 1.20 \\ \text{BP136} = 74150 & \text{Dual 2- to 4-line decoder} & 1.40 & 1.30 & 1.20 \\ \text{BP136} = 74160 & \text{Sync. decade counter} & 1.80 & 1.70 & 1.60 \\ \text{BP136} = 74190 & \text{Sync. blary up-down counter} & 3.50 & 3.25 & 3.00 \\ \text{BP192} = 74192 & \text{Sync. blary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP136} = 74191 & \text{Sync. blary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP136} = 74191 & \text{Sync. blary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP136} = 74191 & \text{Sync. blary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP138} = 74197 & \text{Pre-setable 50MHz beaster} & 1.55 & 5-00 & 400 \\ \text{BP138} = 74198 & 8-118 \\ \text{argue conter} & 1.50 & 1.70 & 1.60 \\ \text{BP138} = 74198 & 8-118 \\ \text{argue conter} & 1.50 & 1.70 & 1.60 \\ \text{BP139} = 74198 & 8-118 \\ \text{argue conter} & 1.50 & 1.70 & 1.60 \\ \text{BP138} = 74198 & 8-118 \\ \text{Argue conter} & 1.50 & 1.70 & 1.60 \\ \text{BP138} = 74198 & 8-118 \\ \text{Argue conter} & 1.50 & 1.70 & 1.60 \\ \text{BP138} = 74198 & 8-118 \\ \text{argue conter} & 1.50 & 1.70 & 1.60 \\ \text{BP138}$		register	0.77	0-74	0.68
$ \begin{array}{c} \text{series} & \text{series} & \text{order} & 0.97 & 9.94 & 0.68 \\ \text{BP105} = 74105 & \text{Single J-K Hip fop equivalent 9001} \\ \text{series} & 0.47 & 0.94 & 0.88 & 0.83 \\ \text{BP110} = 74107 & \text{Dual Master slave flip-flops} & 0.47 & 0.94 & 0.88 \\ \text{BP110} = 74110 & \text{Dual Master slave flip-flops} & 0.45 & 0.45 & 0.50 \\ \text{BP110} = 74111 & \text{Dual data lock-out flip-flop} & 1.25 & 1.15 & 1.00 \\ \text{BP118} = 74118 & \text{Hex set-reset latches} & 1.00 & 0.95 & 0.90 \\ \text{BP118} = 74110 & \text{Hex set-reset latches} & 1.00 & 0.95 & 0.90 \\ \text{BP118} = 74110 & \text{Hex set-reset latches} & 1.00 & 0.47 & 0.84 & 0.58 \\ \text{BP121} = 74121 & \text{Monostable multivibrators} & 0.47 & 0.84 & 0.58 \\ \text{BP145} = 74121 & \text{Monostable multivibrators} & 0.47 & 0.84 & 0.58 \\ \text{BP145} = 74153 & \text{BCD-to-decimal decoder/driver} & 0.47 & 0.84 & 0.58 \\ \text{BP150} = 74150 & \text{D-to-decimal decoder/driver} & 0.47 & 0.48 & 0.58 \\ \text{BP150} = 74153 & \text{D-to-decimal decoder/driver} & 1.60 & 1.70 & 1.60 \\ \text{BP153} = 74153 & \text{D-to-decimal decoder/driver} & 1.60 & 1.70 & 1.60 \\ \text{BP153} = 74155 & \text{D-to-decimal decoder} & 1.50 & 1.70 & 1.60 \\ \text{BP153} = 74155 & \text{D-to-decimal decoder} & 1.50 & 1.70 & 1.60 \\ \text{BP156} = 74155 & \text{D-to-decimal decoder} & 1.50 & 1.70 & 1.60 \\ \text{BP156} = 74155 & \text{D-to-decimal decoder} & 1.50 & 1.70 & 1.60 \\ \text{BP166} = 74160 & \text{Spin. decode conter} & 1.80 & 1.70 & 1.60 \\ \text{BP169} = 74160 & \text{Spin. decode conter} & 3.50 & 3.25 & 3.00 \\ \text{BP199} = 74190 & \text{Sync. up-down decade counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74190 & \text{Sync. up-down decade counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74190 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74190 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74190 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74199 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74199 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74199 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{BP199} = 74199 & Sync. bina$		8-bit bistable latches	1.75	1.65	
$ \begin{array}{c} \text{BP105} = 74105 & \text{Single J-K flip forp equivalent 9001} & \text{Single J-K flip for equivalent 9001} & Single J-K flip for equivalent 900$	BP104 = 74104	Single J-K flip-flop equivalent 9000			
series	PP105 - 7/105	Series	0.97	9-94	0-88
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	DI 100 - 14100		0.07		
$ \begin{array}{c} \text{Br110} = 7410 & \text{Ualleg matter save inp-flops} & 0.55 & 0.56 & 0.56 \\ \text{Br110} = 74110 & \text{Ualleg matter save inp-flop} & 0.25 & 1.5 & 1.00 \\ \text{Br118} = 74118 & \text{Hex set-reset latches} & 1.00 & 0.95 & 0.90 \\ \text{Br118} = 74118 & \text{Hex set-reset latches} & 1.00 & 0.95 & 0.90 \\ \text{Br118} = 74118 & \text{Hex set-reset latches} & 1.00 & 0.95 & 0.90 \\ \text{Br118} = 74112 & \text{Monestable multivativative} & 0.47 & 0.48 & 0.58 \\ \text{Br121} = 74121 & \text{Monestable multivative} & 0.47 & 0.48 & 0.58 \\ \text{Br141} = 74141 & \text{BCD-to-decimal decoder/driver} & 0.47 & 0.46 & 0.58 \\ \text{Br140} = 74145 & \text{BCD-to-decimal decoder/driver} (/C & 1.50 & 1.60 & 0.90 \\ \text{Br150} = 74150 & 16-bit data selector (with stroke) & 1.80 & 1.70 & 1.60 \\ \text{Br153} = 74153 & \text{Dual 4-line-to-1-line data} & 1.20 & 1.10 & 0.95 \\ \text{Br154} = 74155 & \text{Dual 2- to 4-line decoder} & 1.40 & 1.30 & 1.20 \\ \text{Br156} = 74156 & \text{Dual 2- to 4-line decoder} & 1.40 & 1.30 & 1.20 \\ \text{Br160} = 74160 & \text{Sync. decade counter} & 1.80 & 1.70 & 1.60 \\ \text{Br191} = 74190 & \text{Sync. up-down RCD counter} & 3.50 & 3.25 & 3.00 \\ \text{Br192} = 74192 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{Br193} = 74191 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{Br193} = 74192 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{Br194} = 74192 & \text{Sync. binary up-down counter} & 3.50 & 3.25 & 3.00 \\ \text{Br195} = 74192 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{Br198} = 74192 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{Br198} = 74192 & \text{Sync. binary up-down counter} & 1.80 & 1.70 & 1.60 \\ \text{Br198} = 74197 & \text{Pre-setable 50MHz decade counter} & 1.80 & 1.70 & 1.60 \\ \text{Br198} = 74198 & 8-bit parallel L-R shift register & 5.50 & 5-00 & 4-00 \\ \text{Br198} = 74198 & 8-bit parallel L-R shift register & 5.50 & 5-00 & 4-00 \\ \text{Devices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Streis only). \\ \end{array}$	BP107 = 74107	Dual Master size fin-flone			
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	BP110 = 74110		0.55	0.52	0.50
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	BP111 = 74111	Dual data lock-out flip-flop	1.25		
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	BP118 = 74118	Hex set-reset latches	1.00	0.95	
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	BP119 = 74119 PP101 = 74101	Hex set-reset latches. 24-pin	1-85		1.10
$\begin{array}{c} \text{BP150} = 74150 & 16\text{-bit} data selector$	BP141 = 74121 BP141 = 74141	BCD to desimal decider (driver	0.67		0.58
$\begin{array}{c} \text{BP150} = 74150 & 16\text{-bit} data selector$	BP145 = 74145	BCD-to-decimal decoder/driver O/C			
BP131 = 74151 8-bit data selectors (with stroke) 1-00 0-95 0-95 BP135 = 74153 Dual 4-line (-b-1-line data 1-20 1-10 0-95 BP155 = 74154 4-to 16-line decoder 1-80 1-70 1-80 BP155 = 74155 Dual 2- to 4-line decoder 1-40 1-80 1-20 BP165 = 74156 Dual 2- to 4-line decoder 1-40 1-80 1-20 BP165 = 74166 Spic. decade counter 1-80 1-70 1-60 BP191 = 74161 Spic. decade counter 1-80 1-70 1-60 BP191 = 74190 Spic. up-down RCD counter 3-50 3-25 3-00 BP193 = 74190 Spic. up-down RCD counter 2-10 1-95 1-75 BP193 = 74193 Spic. up-down decade counter 2-10 1-95 1-75 BP193 = 74193 Spic. Up-down lecade counter 2-10 1-95 1-75 BP193 = 74193 Spic. Up-down lecade counter 2-10 1-95 1-75 BP193 = 74196 Pre-setable 50MHz decade counter 1-95 1-75 BP194 = 74196 Pre-setable 50MHz decade counter 1-80	BP150 = 74150			1.40	
BP155 74153 Joba 4-line-dora-line data 1-20 0-085 BP155 74154 4-to 16-line decoder 1-30 1-70 1-80 BP155 74155 Dual 2-to 4-line decoder 1-40 1-80 1-20 BP156 74155 Dual 2-to 4-line decoder 1-40 1-80 1-20 BP160 74165 Dual 2-to 4-line decoder 1-80 1-70 1-60 BP160 74161 Sync. decade counter 1-80 1-70 1-60 BP1910 74161 Sync. decade counter 3-50 3-25 3-00 BP1910 74191 Sync. up-down ACD counter (single 3-50 3-25 3-00 BP192 74192 Sync. Up-down decade counter 2-10 1-95 1-75 BP193 74193 Sync. binary up-down counter (tow 2-10 1-95 1-75 BP194 74195 Pre-setable 50MHz decade counter 2-10 1-95 1-75 BP195 74197 Pre-setable 50MHz binary counter 1-80 1-70 1-60 BP195 74197 Pre-setable 50MHz binary count	BP151 = 74151	8-bit data selectors (with stroke)			
BP154 = 74154 4- to 16-line decoder 1-80 1-70 1-80 BP155 = 74155 Dual 2- to 4-line decoder 1-40 1-30 1-80 BP165 = 74155 Dual 2- to 4-line decoder 1-40 1-80 1-70 1-80 BP165 = 74156 Dual 2- to 4-line decoder 1-40 1-80 1-70 1-80 BP160 = 74160 Sync. decade counter 1-80 1-70 1-60 BP190 = 74161 Sync. decade counter 1-80 1-70 1-60 BP191 = 74190 Sync. up-down BCD counter 3-50 3-25 3-00 BP192 = 74190 Sync. binary up-down counter (isrgle 3-50 3-25 3-00 BP193 = 74193 Sync. binary up-down counter (tow 2-10 1-95 1-75 BP193 = 74197 Pre-setable 50MHz decade counter 2-10 1-95 1-75 BP194 = 74196 Pre-setable 50MHz decade counter 1-80 1-70 1-60 BP198 = 74196 Pre-setable 50MHz decade counter 1-80 1-70 1-60 BP198 = 74196 Pre-setable 50MHz decade counter 1-80 1-70 1-60 BP198	BP153 = 74153	Dual 4-line-to-1-line data		1.10	
BP156 = 74156 Dual 2- to 4-line decoder 0/C 1.40 1.80 1.20 BP160 = 74160 Sync. decade counter 1.80 1.70 1.60 BP190 = 74160 Sync. decade counter 1.80 1.70 1.60 BP190 = 74160 Sync. decade counter 1.80 1.70 1.60 BP190 = 74190 Sync. up-down BCD counter 3.50 3.25 3.00 BP191 = 74191 Sync. binary up-down counter (ingle 3.50 3.25 3.00 BP193 = 74192 Sync. binary up-down counter (ingle 2.10 1.95 1.75 BP193 = 74195 Sync. binary up-down counter (ingle 2.10 1.95 1.75 BP193 = 74196 Pre-setable 50MHz decade counter 1.80 1.70 1.60 BP194 = 74196 Pre-setable 50MHz decade counter 1.80 1.70 1.60 BP198 = 74196 Pre-setable 50MHz decade counter 1.80 1.70 1.60 BP198 = 74196 R-bit parallel L-R shift register 5.50 5.00 4.00 BP198 = 74198 B-bit parallel L-Cess shift register 5.50 5.00 4.00 BP198 = 74198	BP154 = 74154	4- to 16-line decoder		1.70	
BP130 = 74166 Dual 2- to 4-line decoder ()(C 1.40 1.80 1.20 BP180 = 74166 Sync. decade counter 1.80 1.70 1.60 BP180 = 74166 Sync. decade counter 1.80 1.70 1.60 BP190 = 74190 Sync. up-down RCD counter 3.50 3.25 3.00 BP191 = 74191 Sync. binary up-down counter (single clock line) 3.50 3.25 3.00 BP192 = 74192 Sync. binary up-down counter (single clock line) 3.50 3.25 3.00 BP193 = 74193 Sync. binary up-down counter (single clock line) 3.50 3.25 3.00 BP193 = 74195 Sync. binary up-down counter (single clock line) 3.50 3.25 3.00 BP193 = 74195 Sync. binary up-down counter (single clock line) 3.50 3.25 3.00 BP193 = 74196 Pre-setable 50MHz decade counter 1.80 1.70 1.60 BP198 = 74197 Pre-setable 50MHz decade counter 1.80 1.70 1.60 BP198 = 74198 8-bit parallel L-R abinary counter 5.50 5.00 4.00 BP199 = 74198 8-bit parallel L-R abinary ceiter 5.50	BP155 = 74155	Dual 2- to 4-line decoder			
$ \begin{array}{c} \mathbf{P100} = 74161 \\ \mathbf{P100} = 74161 \\ \mathbf{Sync. 4-bit binary counter} \\ \mathbf{P190} = 74160 \\ \mathbf{Sync. binary op-down & Counter} \\ \mathbf{Sync. binary op-down counter} \\ Sync. bina$		Dual 2- to 4-line decoder O/C			
BP190 = 74190 Sync. up-down BCD counter 3.50 3.25 3.00 BP191 = 74191 Sync. binary up-down counter (single 3.50 3.25 3.00 BP192 = 74192 Sync. up-down decade counter (single 3.50 3.25 3.00 BP193 = 74192 Sync. up-down decade counter (single 3.210 1.95 1.75 BP196 = 74195 Sync. binary up-down counter (low 2.10 1.95 1.75 BP196 = 74196 Pre-setable 50MHz decade counter 1.80 1.70 1.60 BP197 = 77: setable 50MHz binary counter 1.80 1.70 1.60 BP198 = 74197 Pre-setable 50MHz decade counter 5.50 5-00 4.00 BP198 = 74198 8-bit parallel L-R shift register 5.50 5-00 4.00 BP199 = 74198 8-bit parallel for quantity price. Larger quantities—prices on application (TTL 74 Series only). - - - - - - - - - - - 0 0 - - - - - 0 - 0				1.70	1.60
BP191 = 74191 Sync. binary up-down counter (single clock lne) 3-50 3-25 3-00 BP192 = 74192 Sync. up-down decade counter 2-10 1-95 1-75 BP193 = 74193 Sync. binary up-down counter (town clock lines) 2-10 1-95 1-75 BP196 = 74196 Pre-sctable 50MHz decade counter 1-80 1-70 1-60 BP198 = 74197 Pre-sctable 50MHz decade counter 1-80 1-70 1-60 BP198 = 74197 Pre-sctable 50MHz decade counter 1-80 1-70 1-60 BP198 = 74198 S-bit parallel L-R shift register 5-50 5-00 4-00 BP199 = 74198 S-bit parallel access shift register 5-50 5-00 4-00 D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only). -prices only -prices only	BP101 = 74101 BP100 = 74100	Sync. 4-bit binary counter			1.60
BP192 = 74192 Sync. up-down decade counter 3-50 3-25 3-00 BP193 = 74193 Sync. up-down counter (conter 2-10 1-95 1-75 BP196 = 74193 Sync. bluary up-down counter (conter 2-10 1-95 1-75 BP196 = 74195 Pre-setable 50MHz decade counter 1-80 1-70 1-60 BP198 = 74197 Pre-setable 50MHz decade counter 1-80 1-70 1-60 BP198 = 74198 S-bit parallel Laccess shift register 5-50 5-00 4-00 BP199 = 74199 8-bit parallel access shift register 5-50 5-00 4-00 D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only). 5-50 5-00 4-00	BP191 == 74190	Sync. up-down BCD counter	3-50	3-25	3.00
BP192 = 74192 Sync. up-down decade counter 2:10 1:95 1:75 BP193 = 74193 Sync. blarz up-down counter (tow 2:10 1:95 1:75 BP195 = 74196 Pre-sctable 50MHz decade counter 1:80 1:70 1:60 BP196 = 74196 Pre-sctable 50MHz decade counter 1:80 1:70 1:60 BP198 = 74197 Pre-sctable 50MHz decade counter 1:80 1:70 1:60 BP198 = 74198 sbit parallel 1-R abift register 5:50 5:00 4:00 BP199 = 74198 sbit parallel access shift register 5:50 5:00 4:00 D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only).		clock line)	9.50	9.01	
$ \begin{array}{c} clock lines) . & 2.10 & 1.95 & 1.75 \\ BP196 = 74196 & Pre-scable 50MHz decade counter . 1.80 & 1.70 & 1.60 \\ BP198 = 74197 & Pre-scable 50MHz binary counter . 1.80 & 1.70 & 1.60 \\ BP198 = 74198 & bit parallel L-R shift register . 5.50 & 5.00 & 4.00 \\ BP199 = 74199 & 8-bit parallel access shift register . 5.50 & 5.00 & 4.00 \\ D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only). \\ \end{array}$	BP192 = 74192	Sync. up-down decade counter	2.10	3.25	3.00
$ \begin{array}{c} clock lines) . & 2.10 & 1.95 & 1.75 \\ BP196 = 74196 & Pre-scable 50MHz decade counter . 1.80 & 1.70 & 1.60 \\ BP198 = 74197 & Pre-scable 50MHz binary counter . 1.80 & 1.70 & 1.60 \\ BP198 = 74198 & bit parallel L-R shift register . 5.50 & 5.00 & 4.00 \\ BP199 = 74199 & 8-bit parallel access shift register . 5.50 & 5.00 & 4.00 \\ D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only). \\ \end{array}$	BP193 = 74193	Sync. bluary up-down counter (tow	2.10	1.99	1.10
Br195 = 74195 Pre-setable 50MHz decade counter . 1.80 1.70 1.60 Br197 = 74197 Pre-setable 50MHz binary counter . 1.80 1.70 1.60 Br198 = 74198 8-bit parallel L-R shift register . 5.50 5-00 4.00 Br199 = 74198 8-bit parallel access shift register . 5.50 5-00 4.00 D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only).		clock lines)	2.10	1.95	1.75
BP198 = 74198 8-bit parallel L-R shift register 5.50 5-00 4-00 BP199 = 74199 8-bit parallel access shift register 5.50 5-00 4-00 D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only).	BP196 = 74196	Pre-setable 50MHz decade counter	1.80	1.70	
Br199 = 74198 8-bit parallel L-R shift register 5-50 5-00 4-00 Br199 = 74199 8-bit parallel access shift register 5-50 5-00 4-00 D evices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only).	BP197 = 74197	Pre-setable 50MHz binary counter	1.80	1.70	1-60
D evices may be mixed to qualify for quantity price. Larger quantities-prices on application (TTL 74 Series only).	BF198 = 74198	8-bit parallel L-R shift register	5.50	5-00	4.00
Devices may be mixed to qualify for quantity price. Larger quantities—prices on application (TTL 74 Series only).		o-on parallel access shift register		5.00	4.00
application (TTL 74 Series only).	Devices may be n	nixed to qualify for quantity price. Larg	er quant	tities-pri	ces on
	application (TTL 7	(4 Series only).			

Data is available for the above series of I.C's in booklet form. Price 13p.

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PAL NO.	PAK No.	PAK No.
$UIC00 = 12 \times 7400N$	50p UIC42 = 5×74	50N 50p UIC80 = 5 × 7480N 50p
$UIC01 = 12 \times 7401N$	50p UIC50 = 12 × 74	50N 50p UIC82 = 5 × 7482N 50p
$UIC02 = 12 \times 7402N$	50p UIC51 = 12 x 74	
$UIC03 = 12 \times 7403N$	50p UIC60 = 12 × 74	
$UIC04 = 12 \times 7404N$	50p UIC70 = 8 × 74	
$UIC05 = 12 \times 7495N$	50p UIC72 = 8 × 74	
$UIC10 = 12 \times 7410N$	50p UIC73 = 8 x 74	
$UIC20 = 12 \times 7420N$	50p UIC74 = 8 × 74	74N 50p UIC94 = 5×7494 N 50p
$UIC40 = 12 \times 7440N$	50p UIC75 = 8 x 74	
$UIC41 = 5 \times 7441AN$	50p UIC76 = 8 × 747	76N 50p UIC96 = 5 × 7496N 50p

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15ohm load.

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- some applications. Signal to Noise Ratio 86dB.
- •
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LOGIC Type	DTL 93	D SE	RIES			Price	/C's
No.	Funct	ion			1-24	25-99	100 up
BP930	Expandable dual 4-input 1	VAND			12p	110	100
BP932	Expandable dual 4-input]	AND	huffer		13p	120	11p
BP933	Dual 4-input expander				130	120	110
BP935	Expandable Hex Inverter				13p	120	11p
BP936	Hex Inverter				130		
BP944	Dual 4-input NAND expan			(thous	Tab	12p	11p
DI 011			ouner w				
D.D.C.F					18p	12p	11p
BP945	Master-slave JK or RS				25p	24p	22p
BP946	Quad, 2-input NAND.	du	1000		12p	110	10p
BP948	Master-slave JK or RS	1.00	2010		25p	24p	22p
BP951	Monostable				65p	60p	55p
BP962	Triple 3-input NAND				120	110	10p
BP9093	Dual Master-slave JK with		ate alon	1	400	38p	35p
BP9094	Dual Master-slave JK with	a ocpar	ate cloc	A.	40p		
BP9097	Dual Master-slave JK with	gepan	are croc.	ĸ		38p	35p
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BF3033	Dual Master-slave JK Con	imon C	lock		40p	88p	85p
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BP 701C-SL70IC	TO-5	8	OP Amp	680	50p	45p	
BP 702C-SL702C -	TO-5	8	OP Amp Direct OP	630	50p	45p	
BP 702-72702	D.I.L.	14	G.P. OP Amp (Wide				
			Band)	53p	45p	40p	
BP 709	D.I.L.	14	High OP Amp	53p	450	400	
BP 709P-µA709C	TO-5	8	High Gain OP Amp	53p	45p	400	
BP 711-µA711	ŤŎ- 5	10	Dual comparator	58p	50p	450	
BP 741 -72741	D.I.L.	14	High Gain OP Amp	UUU	Sop	Hop	
			(Protected)	75p	60p	50p	
µA 703C-µA703C	TO-5	6	R.FJ.F. Amp	43p	350	275	
TAA 263-	TO-72	4	A.F. Amp				
TAA 293-				70p	60p	55p	
(AA 293	TO-74	10	G.P. Amp	90p	75p	70p	



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VOL 47 NO 6

Issue 776

OCTOBER 1971

Who makes what?

WELL, of course, that's easy! When you buy a radio set, hi-fi amplifier or piece of electrical equipment, the trade mark tells you that. Or if the mark is an unfamiliar one you can find the piece moulded on the cabinet or printed on the carton which says 'Made in Outer Mongolia' or whatever.

PRACTICAL

People are interested in where products are made for a variety of reasons. The possible standard of workmanship, value-for-money, reputation, reliability, the position of after-sales service, availability of spare parts—or even personal political motives—are all valid reasons why a purchaser would want to know who makes what, and where.

From November 30th, however, the Merchandise Marks Act comes to an end and from that date it will no longer be obligatory for imported goods to bear a 'country-oforigin' mark. The casual customer will therefore not be sure if he is buying a radio set made in Birmingham or Baluchistan, or an amplifier which is a Chinese-copy of an American design made in Hong Kong or Japan.

Of course, there will still be trade marks which proclaim that the goods are true-blue British or come from wellrespected foreign companies. Sadly, though, things are not always what they seem for many tape recorders, radio and TV receivers, bearing the names of old-established British manufacturers have for some time been made overseas and imported for the home market.

From November 30th, protection of the consumer has been thown at the retailer who, when asked, will be expected to disclose where a particular product is made. Unfortunately in may cases he will have no way of giving the required information. For instance, 'household-name' UK companies are none too keen to admit that some of their goods are being made in low-cost labour countries. Moreover, if a retailer does not know the country of origin, or (even inadvertently) gives the wrong answer, he will be at risk under the Trade Descriptions Act!

Apart from the factors mentioned earlier, which allow a customer to assess the advisability or otherwise of making a purchase, a particularly disturbing aspect of electrical goods (notably electric blankets and appliances) is that safety specifications may fall far short of those imposed on British-made goods and expose the user to potential hazards.

So, together with the British Radio Equipment Manufacturers' Association, the Radio and Jelevision Retailers' Association and the Association of Manufacturers of Domestic Appliances, we beg the Secretary for Trade and Industry to think again. And to would-be purchasers after November 30th we can only say, most emphatically *caveat emptor!*

W. N. STEVENS-Editor.

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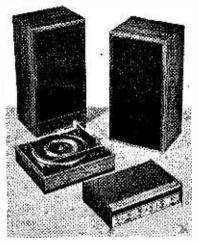
FREE WALL CHART: Circuit Building Blocks

NOVEMBER ISSUE WILL BE PUBLISHED ON OCTOBER 8th

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Viscount value



The Viscount III Audio System is the first of a range of audio equipment marketed in the UK by R+TV Components (Acton) Ltd. under the Viscount brand name.

The complete system includes a choice of two amplifiers, a Garrard SP25 Mk III turntable and two Duo type II or type III speakers.

Amplifier RT100 is for ceramic cartridges and model RT101 which has an additional input stage accepts magnetic types. B ot h amplifiers incorporate f.e.t.'s giving excellent signal to noise ratios.

Prices are: RT100, £17; RT101, £22; Garrard SP25 with ceramic cartridge, £21 or with magnetic cartridge, £23. Duo type II speakers cost £14 the pair and Duo type III, £32.

Savings can be made by buying complete systems.

Impex cases

In last month's issue we stated the following about Impex instrument cases:

"At no additional cost, Impex cases can be supplied with all or some of these specifications: alternative colours of blue or bronze; no hood; no ventilating slots or back/front anodised front/rear panels".

What we should have mentioned is that these additions can only be supplied free of charge for quantities over 24.

NHK's anniversary

June 1 saw the anniversary of the Radio Japan Service. Thirty-six years ago NHK commenced this service in commemoration of the 10th anniversary of radio broadcasting in Japan and the attainment of 2 million licensed subscribers.

From June 1934 to 1945, NHK broadcast its domestic programmes to Korea, Taiwan and Manchuria, then Japan's outlying territories, utilising telephone channels of the International Telephone Company. Shortwave broadcasts were received in various parts other than the areas directed, and reception reports began to come to NHK from Japanese residing abroad and from large numbers of overseas people. In those days, Japanese living in various parts of the world totalled nearly two million. These people raised their voices requesting the establishment of an overseas broadcasting service. In Japan, too, commencement of an overseas broadcasting service was also strongly desired, since this country had been isolated after her withdrawal from the League of Nations and was in need of reinforcing its public relations abroad. At first the service was known as Radio Tokyo (Radio Japan was first used in 1952).

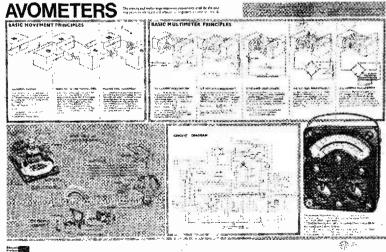
Avo wall chart

Avo Limited (Thorn Group) announce a new wall chart featuring the latest Mark 4 version of the Avometer Model 8. The design and layout is such that it will prove of great interest not only to Avometer users but also to technical schools and general education establishments as a teaching aid.

Information is presented in a clear, logical manner and the theoretical and practical information is both concise and easy to understand. Magnetic theory is developed pictorially to outline the operation of the moving coil movement whilst a mathematical treatment is used to derive the various ranges of a multimeter.

Illustrations of the interior of the instrument include an exploded view of the movement to simplify the constructional detail and a circuit diagram provides an introduction to standard symbols and references.

The Avometer chart (size 20×30 in., 508×762 mm.) is ideal for blackboard or wall presentation and is available from Spares Dept., Avo Limited, Avocet House, Dover, at 75p per copy, including post and packing.



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Windermere VHF

The v.h.f. radio services from the Windermere relay station started on August 7. Radio 2 is on $88 \cdot 6MHz$, Radio 3 on $90 \cdot 8MHz$ and Radio 4 on $93 \cdot 0MHz$, all with horizontal polarization.

The v.h.f. radio transmissions from this new relay station, which is situated on Claife Heights, above the western shore of Lake Windermere, will serve Ambleside, Windermere town, Bowness-on-Windermere, and an area around the northern shores of the lake.

Stereo from Rowbridge

The necessary equipment has now been installed at Rowridge to enable the Radio 3 transmitter (90.7MHz) to broadcast stereophony. This is being brought into use immediately, so that listeners can hear the Promenade Concerts in stereo-all except four of the concerts will be broadcast stereophonically in Radio 3. The stereo transmissions from Rowridge will initially be on an engineering test basis; every effort will be made to maintain them, but there may be occasions when it will be necessary to revert to monophonic transmissions. The date for the start of the full stereophonic service will be announced later.

Miniature Electric Tools

A range of battery-operated miniature tools has been introduced by Expo (Drills) Ltd., 62 Neal Street, London, W.C.2.

Two basic models are available. The 'Reliant' is designed for lighter work such as model making, and has a full load current of 1.5A. Its rated torque is 1.38oz. in. (100 gm.cm.). For jobs requiring a more powerful tool and for professional applications the 'Titan Super' is rated at 3.5A on full load. It has a rated torque of 350cmp operating at 4000-9000rpm. An extremely large range of accessories is also available for both models. Various collets and accessories are supplied with the different models, which range in price from £3 to £5.50.

Lectures

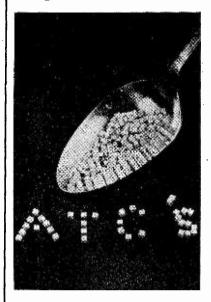
THE Polytechnic of North London Department of Electronic and Communications Engineering is running a course on Audio and Acoustic Measurements on Thursdays 6.30-8.30 p.m. 1971/2. 1971: Oct., 28, Parameters and Standards of Measurement R. N. Baldock, B.Sc., F.B.K.S.T.S. Nov. 4, Measurement of Microphone Performance, H. D. Harwood, B.Sc., B.B.C. Research: 11. Electrical and Mechanical Measurements in Tape Recording, R. L. West, B.Sc., C.Eng., F.I.E.R.E.; 18, Measurements in Professional Recording, Angus McKensie; 25, Electrical and Mechanical Measurements on Disc. Pickups and Accelerometers, S. Kelly, C.Eng., F.I.E.R.E. Dec. 2, Measurement of Loudspeaker Performance I, R. L. West; 9, Measurement of Loudspeaker Performance II, R. C. Driscoll, M.Sc.: 16. Measurement of Studio Characteristics, R. C. Driscoll, M.Sc. 1972: Jan. 13, Measurement of Noise Transmission, J. Moir C.Eng., F.I.E.E.; 20, Assessment of Noise Nuisance, J. Moir, C.Eng., F.I.E.E.; 27, Visit to A.I.R.O. Laboratories, Hemel Hempstead, G. Berry, B.Sc., C.Eng., M.I.Mech.E. and Assocs.

Grand Junk Sale

The Star Short Wave Club, G3ZWA, is holding a "Grand Junk Sale" in aid of Radio Amateur Invalid Bedfast Club funds. It is to be held at the New Inn Hotel, Bramley Town Street, Bramley, Leeds 13. The date is Wednesday September 15th and time 7.30.

If any readers have any junk, the Star Club will collect within a radius of 25 miles from Leeds. Please contact Mr. T. Leeman, G8BUU, 115 Asket Drive, Seacroft, Leeds, LS14 1HX.

Chip "C's"



Featuring a very low insertion loss, a very high Q and high capacitance-to-volume ratios, a new range of ceramic and porcelain chip capacitors from Guest International Limited is ideal for u.h.f. and microwave applications. Apart from the basic pellet and chip, there is a wide choice of lead variations: stripline, axial and radial wire, axial and radial ribbon.

ATC 100 series are porcelain chips with an insertion loss of 0.03dB, a Q of over 10,000 and a capacitance range of 0.1 to 1000pF. In a case size of only 0.110in.square, the working voltage is up to 500V d.c.

ATC 200 series are ceramic chips with an insertion loss of 0.05dB. The capacitance range is 10pF to 1.2μ F and chips are available in case sizes 0.050, 0.110 and 0.250in.square, depending on the value and the working voltage.





So many audio sources are now in common use that an article describing a specific design would meet the needs of only a few. This problem is overcome here by giving circuits for virtually all sources and care in design has made possible a large number of permutations so that a custom built audio mixer is possible even for those readers unable to design their own. The clever constructional method employed also allows for rapid modification to meet any mixing need.

A STUDY of the four block diagrams in Fig. 1 will show how the various signal preamplifiers, the tone control unit and line amplifier etc. can be used to build audio signal mixing systems to cater for tape recording, public address, discotheque and even Hi-Fi applications. The combinations given are but a few of the permutations which can be extended for stereo use simply by duplication.

Each signal preamp has the same output level, 100mV, and will match directly into either the tone control unit or the line amplifier, both of which are designed to accept a signal level of 100mV. The tone control has unity gain so this provides 100mV output to match into the line amplifier. Any group of signal preamplifiers can therefore be coupled directly to the line amplifier or via the tone control unit and then to the line amplifier. This allows any of the signal preamplifiers to be used with or without the tone control unit.

The line amplifier has a nominal output of 1V r.m.s. at a fairly low impedance and can be used to drive a power amplifier module requiring between 500mV to 1V input, or it can be coupled to a tape recorder or radio input, of to a public address or music amplifier with inputs rated for 500mV to 1V. Groups of the preamplifiers and line amplifier can be run from batteries or from an 18V mains power supply.

COMBINATIONS

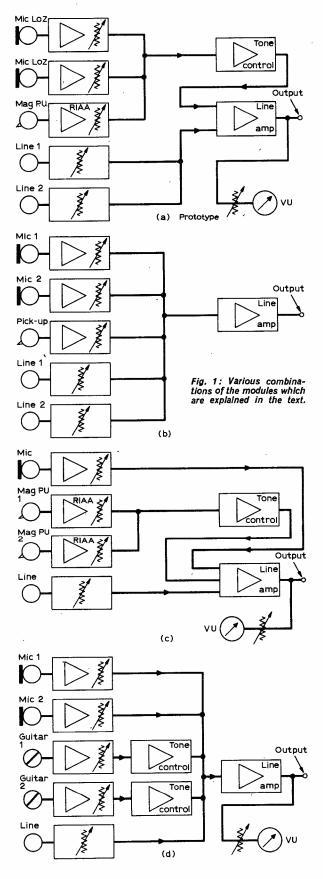
A fairly simple arrangement is shown in Fig. 1a and is intended for use as a general purpose signal mixer as shown on the front cover. This would be suitable for tape recording, as it has two microphone inputs, one input for low impedance pick-up (with R.I.A.A. correction), and two line level (200mV to over 1V) inputs. The tone control operates for the mic. and pickup inputs, whilst the line inputs go directly to the line amplifier. The system also includes a VU meter with variable control so that the meter can be set to read with the level meter on a tape recorder.

The combination shown in Fig. 1b has the same input channels but no tone control unit. The arrangement in Fig. 1c might be suitable for a small discotheque with one mic. input, two pickup inputs both via a tone control unit and one line input, suitable for the output from a tape recorder. For music application, Fig. 1d provides two mic. and two guitar inputs with separate tone control units for each guitar input. In each case the VU meter is optional but can be used as an overload indicator as will be shown later.

PERFORMANCE SPECIFICATIONS

Seven preamplifiers will be described, each catering for specific signal sources as follows:—

- LOW IMPEDANCE MICS. (25 to 600Ω). Direct input, no transformer required. Input sensitivity 50µV. Output 100mV. Frequency response ±1dB. 30 to 30,000Hz. Signal-to-noise -50dB. (Code: Mic. Lo-Z).
- FOR MICROPHONES WITH BUILT-IN TRANS-FORMERS. Medium to high impedance. Input sensitivity 1mV. Output 100mV. Frequency response ±1dB 20 to 30,000Hz. Signal to noise -60dB. (Code: Mic. Hi-ZT).
- FOR CRYSTAL MICS. High impendance. Input sensitivity 2mV. Output 100mV. Frequency response ±1dB 30 to 12,000Hz. Signal-to-noise -50dB. (Code: Mic. Hi-ZX).
- MAGNETIC PICKUP. Impedance 56kΩ. Input sensitivity 5mV. Output 100mV. Frequency response to R.I.A.A. as shown in Fig. 2. Signal-tonoise - 60dB. (Code; PU Mag.)



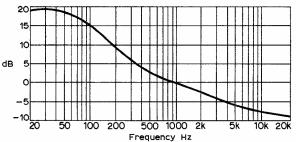


Fig. 2: The RIAA response curve incorporated in the magnetic pickup module.

- CERAMIC PICKUP. Input impedance 2MΩ. Input sensitivity 60mV. Output 100mV. Frequency response approximately 20 to 20,000Hz ±1dB. Equalisation automatic due to high input impedance. Signal-to-noise -50dB. (Code: PU Cer.)
- CRYSTAL PICKUP. Input impedance about 1MΩ. Input sensitivity 850mV. Output 100mV. Frequency response approximately 20 to 20,000Hz ±1dB. Equalisation automatic due to high input impedance. Signal-to-noise -60dB. (Code: PU Xtal.)
- 7. GUITAR PREAMP. Input impedance medium, 10 to $100k\Omega$. Input sensitivity variable. 50 to 100mV. Output 100mV. Frequency response $\pm 1dB$ 20 to 30,000Hz. Signal-to-noise -50 to -60dB. (Code: G.)

The tone control unit will take the output from any of the preamps and provides 100mV output so that it can be coupled directly to the line amplifier. Its performance is as follows:— With tone controls neutral the flat response is 10 to $80,000Hz \pm 1dB$. The response of the bass and treble controls is shown in Fig. 3 providing 15dB lift and 12dB cut at 40Hz and 15dB lift to 12dB cut at 20,000Hz The signal to noise performance is better than -90dB. (Code: **TC**).

The line amplifier unit will accept the output from any of the preamps or the tone control unit and therefore has an input sensitivity of 100mV. The output signal level available is 1V r.m.s at an impedance of approximately 600Ω . The frequency response is flat from 10 to nearly 100,000Hz and the signal-to-noise performance -80dB. The output can be fed simultaneously to one or more tape recorders and/or amplifiers via series output resistors as will be shown. The overall gain of the line amplifier is adjustable i.e., the input sensitivity can be increased. (Code: LA.)

The line inputs do not require separate preamplifiers as these are arranged in passive network coupling directly to the line amplifier. The input sensitivity will be 200mV to over 1V r.m.s. at approximately $10k\Omega$ impedance. (Code: Linput.)

CONSTRUCTION

The construction of each of the preamplifier modules is such that the circuit board is mounted on a small screen this, in turn, being secured to the fixing bush of the gain control (VRI in each circuit). The complete preamp can therefore be secured to a front panel by the lock nut of the gain control. The tone control unit is similar except that there are two potentiometers and both are mounted on the small screen and then secured to the panel. The line amplifier can be simply mounted on a piece of aluminium angle (${}^{3}_{8} \times {}^{3}_{8}$ in) or on stand-off pillars as convenient. It need not be fixed to a panel but must

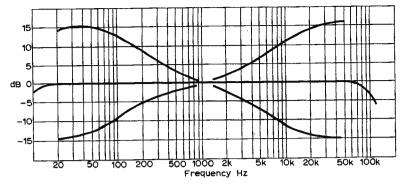


Fig. 3: Response curves of the tone control module.

be fairly well clear of mains power supply components. The input sockets for all inputs must be insulated from the front panel and earthed only at the common earthing point on the preamplifier board. If the mixing unit is to be mains powered, the components for the power supply must be kept as far away from the amplifier boards as possible and preferably with a screen around them.

The seven preamplifier boards are exactly the same and details for construction are given in Fig. 4. In each case the gain control (VR1) is fitted as shown

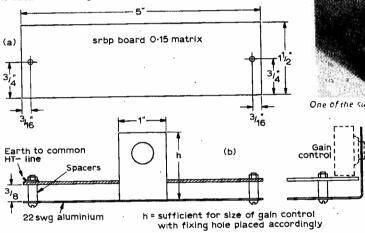
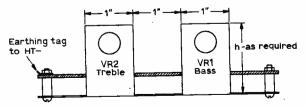


Fig. 4 : Details of the sub-chassis used for each module.

in Fig. 4c. The construction of the tone control board is similar except that two controls have to be fitted so the screen has two flaps as shown in Fig. 5. The height of each flap and the position of the fixing hole will depend on the physical size of the potentiometers used.

CIRCUITS AND ASSEMBLY OF THE MODULES

Each of the amplifier modules and the tone control



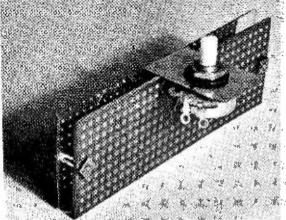
Controls mounted as in Fig.4 (c)

Fig. 5: Modified sub-chassis for the tone control module.

can be tested separately by operating them from an 18V battery (two 9V units) and by checking each one with an audio signal generator and audio voltmeter or by connecting an appropriate input signal source and feeding the output to an amplifier.

LINE AMPLIFIER

The best order of construction is to build the line amplifier (code: LA) first so that the preamps can be tested with it as each is completed. The circuit is given in Fig. 6 and the board layout in Fig. 7. As

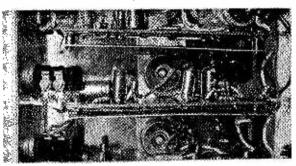


One of the sub-chassis before the components are fitted; dimensions are given in Fig. 4.

(c)

with all the circuits except the tone control unit only one type of transistor is used—the BC109 silicon NPN—which is readily available and inexpensive. The line amplifier employs three of these beginning with Tr1 as an emitter follower input stage with Tr2 as the amplifier.

Negative feedback between Tr2 and Tr1 via the pre-set PR1 ensures a linear response and some adjustment over total gain. The pre-set should be adjusted so that with a 100mV input signal the output at C6 is exactly IV. If a signal generator and audio voltmeter are unavailable, set PR1 to midway position. The output transistor Tr3 operates as an emitter follower and provides a low impedance



An internal view of the prototype showing one of the modules. The input jack is on the left with the common wiring on the right.

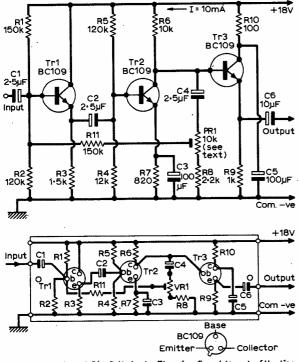


Fig. 6: (above) and Fig. 7 (below): The circuit and layout of the line amplifier module.

output. Details for feeding more than one power amplifier and/or tape recorder from the line amplifier output will be given later.

MICROPHONE PREAMP (Mic Lo-Z)

The circuit is given in Fig. 8 and the board layout

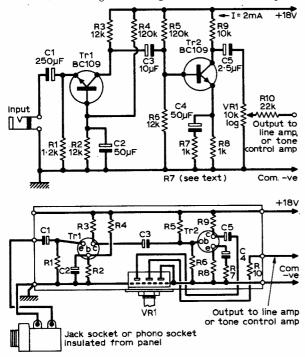


Fig. 8 (above) and Fig. 9 (below): The circuit and layout of the low impedance microphone preamplifier module.

★ components list

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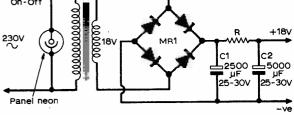


Fig. 10 : Suggested circuit for a mains power supply to power the mixer.

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NEW 1000kW station on 1457kHz located near Tirana in Albania is being heard in this country as a background to BBC Radio 4. (South and West). It relays Radio Peking and has been heard at 2100hrs GMT with the identification 'Govarit Pekin' followed by programmes in Russian. An increasing number of superpower broadcasters are appearing on the medium waves. Urumchi in Western China (2000kW) is audible all evening on 1525kHz; listen between Prague 1520kHz and Vatican Radio 1529kHz for Russian speech plus jamming. Libya has two megawatters—El Beida on 1124kHz and Tripoli on 1250kHz. El Beida is the stronger of the two; it is heard with EAJ15 Barcelona on the same frequency throughout the evening. Tripoli suffers from interference from Dublin/Cork until 2245hrs when the latter close down, leaving the channel clear. Further east, Riyadh (1200kW) in Saudi Arabia is on 587kHz and has been logged before midnight with Arabic programming.

Calcutta 1130kHz (1000kW) was heard frequently last winter before sign-off at 1700hrs. It is now on extended schedule and has been several times during the summer at 2230hrs GMT, as it changes from Chinese to Indian programming. There are three, Voice of America stations in the Far East, each with 1000kW, all of which have been heard in the U.K. Bangkok, Thailand 1580kHz was logged at the end of September last year at 2200hrs GMT while Poro 1140kHz in the Phillipines and Okinawa 1178kHz in the Ryuku Islands are both heard in winter before sign-off at 1700hrs.

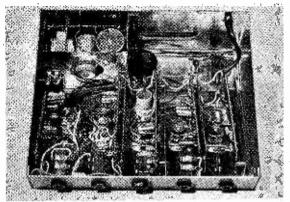
'Book of the World 1971' is the title of an unusual paperback now on sale in the bookshops. Sections deal with 181 different countries, each with a map plus details of the type of government, languages spoken, religions and currency. As well as providing the DXer with a valuable aid to station identification, this book gives background information on countries logged and adds interest to the hobby.

Broadcasts in English on the medium waves can be heard from a number of unexpected places during the late evening. Timisoara, Rumania on 755kHz has a programme in English, French and German at 2230hrs GMT, which should not be confused with the Voice of the West, in Portugal, also on 755, which starts its English programmes at 2245hrs. The Voice of America in Rhodes is 1259kHz is heard regularly in English between 2100hrs and midnight, with some interference from Wroclaw in Poland. Rhodes is one of the Dodecanese Islands situated near the coast of Turkey, which are usually counted as a separate country by DXers. The VOA is also heard in English from Thessalonika, Greece 791kHz between 2100hrs and 2130hrs. Enugu, Nigeria 1320kHz, although only 10kW, is a weak but consistent signal at 2300hrs GMT when it carries it's final news in English before closing down. ELBC Monrovia, Liberia on 629kHz has English programming and is usually logged between 2330hrs and sign-off at 0045hrs.

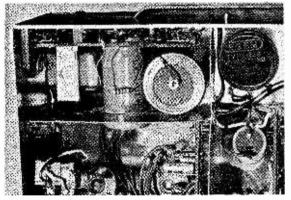
in Fig. 9. The input transistor Tr1 operates as a grounded base amplifier and provides a very low impedance input. Tr2 is a signal amplifier with negative feedback introduced by R7 in series with the emitter by-pass capacitor C4. This resistor can be reduced a little in value to obtain more gain for very low output microphones, although it should not be less than 680Ω otherwise the noise level will rise excessively. As with all the preamplifiers the output is taken to the gain control VR1 and thence via the series resistor (R10) which becomes part of the passive mixing network common to all the preamps.

MAINS POWER SUPPLY

The current consumption of the preamplifiers at



Internal view of the completed prototype.



The power supply should be properly screened as shown.

18V varies between about 2 and 5mA. The line amplifier takes 10mA and the tone control unit 5mA. Four preamplifiers, a line amplifier and say one tone control unit, would require about 25mA which could be drawn from an 18V battery with reasonable economy. Two PP9 batteries would be ideal. Larger mixing units, particularly if built for stereo use and requiring duplication of preamps etc. are best operated from a mains power supply. One that will provide well over 100mA is given in Fig. 10 (see components list for transformer and rectifier details). The resistor R will be around 1k Ω for low current drain and of lower value for higher current drain. With the mixing circuits running, adjust R to obtain 18V across C2.

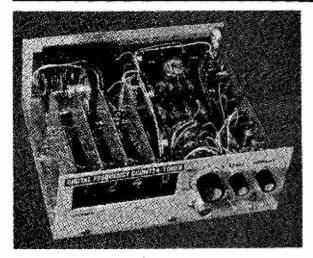
Part 2 next month, will deal with the rest of the preamplifier modules.

TO BE CONTINUED



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DIGITAL FREQUENCY COUNTER/TIMER



PART 2 (continued from the September issue)

This negative going step is passed by C5 to Tr7 base which moves by -5 volts thus cutting off Tr7 and allowing the collector of Tr7 to rise to +5V.

The current through VR2 and R14 discharges C5 so that the negative voltage on Tr7 base rises exponentially until Tr7 conducts. The rate is determined by the setting of VR2, the DISPLAY control.

When Tr7 conducts, its collector voltage starts to fall, turning off Tr6 and thus increasing the base current to Tr7 and producing a fast negative going voltage at Tr7 collector. This voltage step is differen-

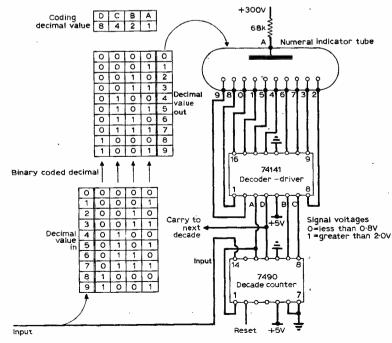


Fig. 9: Table showing the operation of the decade counter and decoder-driver.

J. THORNTON-LAWRENCE GW3JGA

NOTE

Several wiring connections were omitted from figure 6 in the September issue. Pin 10, on IC's 3, 4, 5, 6 and 7, should be shown connected to the OV rail. Pins 1 and 12 on IC's 9, 11, 13 and 15, should be shown joined.

On IC1, pin 4 is the +5V supply rail and pin 11 is OV supply rail. On IC2, pin 14 is +5V supply rail and pin 7 OV supply rail. These latter connections were omitted from figure 6 for clarity.

tiated by C7, R16 and R17 and inverted by IC2d to give a positive pulse of $200\mu S$ duration.

The positive going edge of this pulse resets the counter to zero and the negative going edge sets the JK flip-flop IC1a and causes the cycle of operation to be repeated. The time for which the reading is displayed is controlled by the DISPLAY control and this is adjustable from about 0.5 second to 2 seconds.

Counter and Display

The counter and display has four decades, each decade operates in the same way. The output of one decade is linked to the input of the next higher decade to perform the 'carry' operation and a reset connection is common to all decades.

The input to the counter is a series of pulses. The counting has to be carried out in binary form and then decoded into decimal form for display.

This is done using a decade counter I.C. type SN7490N, which consists of 4 master-slave flip-flops forming a $\div 2$ counter and a $\div 5$ counter.

The input signal goes to the input of the +2 counter (pin 14) and the output of this (pin 12) is linked externally to the input of the $\div 5$ counter (pin 1).

The outputs of the four flip-flops are called A, B, C, D and have a decimal value of 1, 2, 4, 8 respectively. The table showing the operation of the decade counter is included in Fig. 9.

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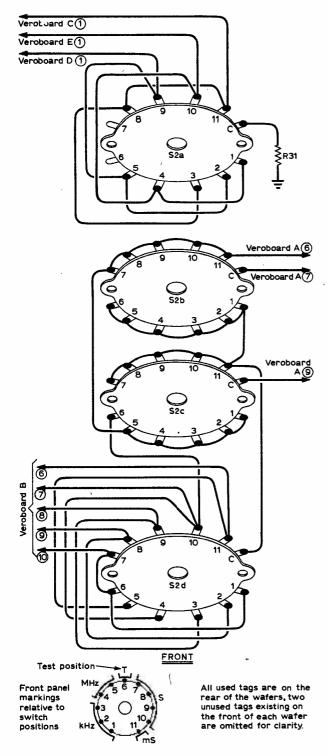


Fig. 10 : Exploded view of the function switch S2, viewed from the front.

The counter can be reset to zero at any time by the application of a positive pulse to pin 2.

The binary coded decimal output from the counter, A, B, C, D is connected directly to the inputs of the decoder/numeral tube driver I.C. type SN74141N. Here the inputs are decoded back into decimal form for driving the numeral tube (Hivac Type XN13).

The numeral tube contains neon gas at low pressure and has an anode and 10 wire cathodes in the form of numerals, 0-9.

With the anode connected to HT + through a current limiting resistor, any cathode which is connected to 0V will be surrounded by a glow showing up the shape of the number.

The outputs of the decoder-driver are collector circuits of high voltage transistors which, when made to conduct, effectively connect the appropriate cathode to 0V. For example, suppose the input to the decade counter receives 5 pulses or cycles of input signal. The output will be D low, C high, B low, A high, representating a binary number of 0101. This number is decoded in the decoder/driver to figure 5 and the figure 5 output (pin 14) becomes switched to 0V causing the figure 5 to be displayed in the numeral tube.

Each counter, decoder/driver and tube is built on a Veroboard panel and the inter connection of these panels is shown in the main diagram Fig. 6.

The decimal indication is provided by miniature wire ended neon indicator lamps mounted between two numeral tubes and these are switched by the function switch S2a.

Function Switch

The function switch S2 is made up from Radiospares Miniature 'Maka Switch' parts and consists of a shafting unit and 4 break-before-make 1 pole 12 way wafers. The wafers are spaced by 2 spacers on each side. Only 11 positions are used and the stop on the indexing mechanism is adjusted accordingly.

The wafers are lettered a, b, c and d, starting at the rear and working towards the indexing mechanism. Wafer 'a' controls the decimal point neon lamps, wafer 'b' the clock period input to the gate, wafer 'c' the signal input to the gate and wafer 'd' the output from the clock divider.

In the frequency positions 1 to 5, the output from the input stage is fed to the 'Signal Input' of the gate and the appropriate clock signal is fed to the 'Clock Period Input' of the gate. In the time positions, 7 to 11 of the switch, the output from the input stage is fed to the 'Clock Period Input' of the gate and the appropriate clock signal is fed to the 'Signal Input' of the gate.

A test position is provided in position 6 where the lkHz clock output is counted for a period of 1 second to give a reading of 1,000. The test position only serves to check that the gate, counter and display are functioning, it does not provide a check of the crystal oscillator frequency.

The wiring and connections to the switch are shown in Fig. 10. The method of assembly is as follows:

1. On the shafting unit, check that the lengths of studding are screwed fully into the indexing mechanism plate and remove the end nuts and washers.

2. Rotate the switch spindle to the fully anti-clockwise position (as viewed from the front). The flats on the wafer end of the spindle should be in line with the studding. This is position "1".

3. The switch should now be held so that the lengths

of studding are at each side and the control knob end of the spindle is towards you, as shown in Fig. 10.

4. Take a 1 pole 12-way wafer and hold this with the side having the majority of tags, away from you and with the two tags that are facing you positioned in the top right hand segment. Rotate the rotor section until the common or rotor contact is making contact with tag 1. The slot in the rotor should now be horizontal.

5. Assemble the wafer on to the shafting assembly, keeping the side with the majority of tags facing away from the indexing mechanism. This wafer is now wafer "d" in Fig. 10.

6. Fit 2 spacers on each piece of studding.

7. Fit another wafer as described in operation 4 and 5 (wafer "c").

8. Repeat operations 6 and 7 until four wafers have been fitted (wafers "b" and "a").

9. Fit washers and fixing nuts to the studding.

10. The rotation of the switch should now be checked to see that all parts operate smoothly.

11. The rotatable stop plate, which is located on the face of the indexing mechanism, should now be set to give 11 positions, working clockwise and commencing from the maximum anti-clockwise position. This is the position at which all the wafers have been set. The stop plate may be held in place temporarily by the mounting nut, ready for assembly to the front panel.

12. The wiring of the switch may now be carried out as shown in Fig. 10. 22 s.w.g. tinned copper wire is used for wiring from tag to tag, sleeving being necessary only when crossing between wafers.

13. The correctness of the wiring can be checked by using an ohm or continuity meter and measuring between the appropriate tags whilst the switch is rotated, e.g. S2b/C is connected to S2d/C in positions 1, 2, 3, 4, 5 and 6.

Power Supply

The power supply consists of a Radiospares Midget Mains transformer which with its associated rectifiers smoothing and regulators provides

+300V	at 10mA	Unregulated
+5V	at 350mA	Regulated
-5V	at 5mA	Regulated

The output from the 250 volt secondary winding is rectified by D7 and smoothed by C12 to give about +300V output. The negative return is through R24 and D9 to 0V. The negative supply derived in this way is smoothed by C13 and regulated by the zener diode D9 to give -5V. The I.C. logic requires about 350mA at +5V and this is provided by the 6.3 volt filament winding on the transformer.

The 6.3 volt output is full wave bridge rectified to give about +7.5 volt across the reservoir capacitor C14. This feeds the zener diode D6 and the power emitter follower Tr11 to give the required 5V output.

Veroboard Panels

The layout drawings of the 0.1×0.1 in matrix Veroboard show the copper strip side of the panel the components on the reverse side being "seen through" the copper strips.

Each copper strip is given a letter code and each

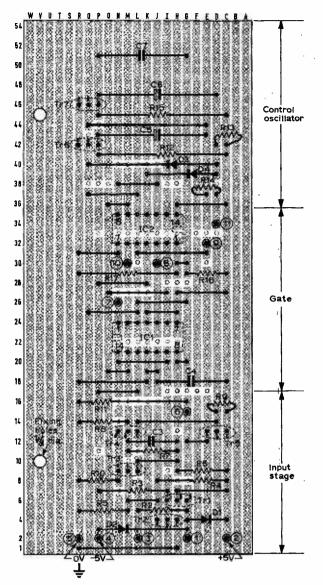


Fig. 11. Veroboard panel A. Input stage, gate and control oscillator. Viewed from copper side. Note correct orientation of integrated circuits when fitting. The integrated circuits and their holders are indented at one end to assist orientation.

horizontal row of holes, a number.

All solid horizontal lines represent wire links which connect one strip to another.

The orientation of integrated circuits and transistors should be noted carefully, some transistors may have unusual lead positions and these may have to be crossed over for the connections to be correct. Assembly of the components, etc., must be carried out in a systematic manner or errors will result. A suggested method is as follows:

1. Cut the Veroboard to the correct dimensions using a fine toothed saw and working from the copper side. Apart from checking the physical dimensions, also check that the correct number of holes appear in each direction.

2. Fix strips of masking tape on all four edges of the Veroboard panels. Also fix masking tape to the same edges on the reverse (copper) side of the board.

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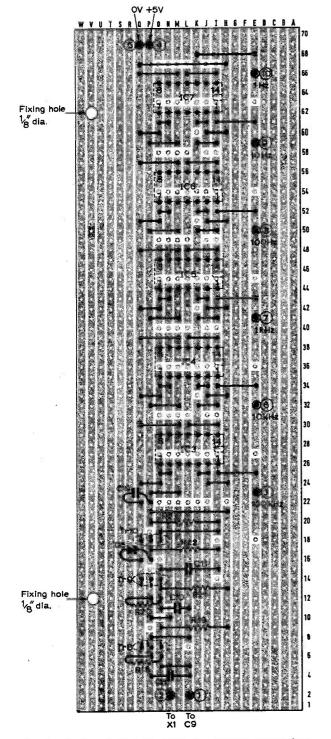


Fig: 12. Veroboard panel B. Crystal clock and divider. Viewed from copper side

3. By writing on the masking tape, letter all the vertical strips and number all the horizontal rows of holes on both sides of the board.

4. The breaks in the copper may be made using the appropriate Vero tool, part No. 2022 or by using a $\frac{1}{16}$ sin. drill held in a pin vice. Each copper break on the layout drawing should be identified by its co-

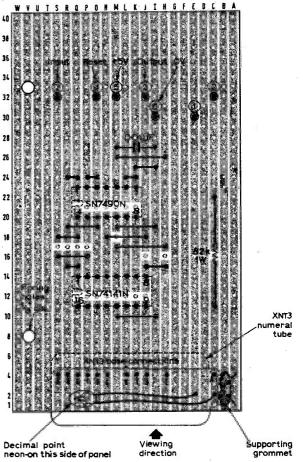


Fig. 13: Counter and display panels C. D. E. F. Panel F. does not require a neon "decimal".

ordinates, e.g. D38 and the copper strip removed by carefully rotating the tool or drill by hand using the identified hole as a guide.

Experience has shown that just counting the position of a hole is not reliable and errors are very likely to result.

5. Wire links are formed from pre-stretched 22 s.w.g. tinned copper wire. A length of about 3ft of 22 s.w.g. TC wire is stretched slightly by using a pair of pliers at each end, until it becomes stiff. All wire links are fitted on the components side.

A length, in excess of the length of the link required, is cut off and two right angled bends are made in the appropriate places to form the required "U" shaped link. A pair of pointed nosed pliers is useful for this and the span of the link can be checked by holding it against the panel in the appropriate place. When the link is inserted, the end wires may be bent outwards slightly to hold the link in place while it is soldered.

 $\overline{6}$. Components should be fixed starting at one end of the panel and the position and value checked before soldering.

7. The soldering should be carried out using a soldering iron with a small chisel shaped bit, no wider than $\frac{1}{16}$ in. The cored solder used should be no larger than 20 s.w.g.

Check each soldered joint using a magnifying glass as it takes only a very small piece of swarf or solder to bridge the 0.025in gap between the copper strips.

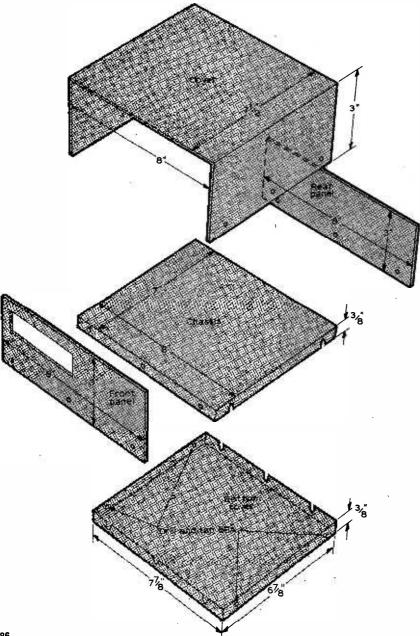
8. Connections to the Veroboard panel are made to pins inserted in the appropriate places and soldered to the copper strip. These pins are Vero type TP 11032 or similar.

An insertion tool for this pin is Vero part No. IT 2151.

9. Tick off each break, link, component and pin on the drawings as they are fitted, this will identify any items which otherwise might be missed.

Semiconductors

The 3702 and 3704 transistors are specified in the components list as being the '2N' types, these are the most popular version and are readily available. The 3702 and 3704 transistors are also available



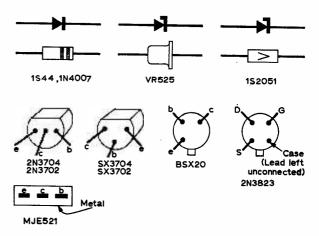


Fig. 14 (left) Cabinet and chassis construction.

in an 'SX' version. The lead connections for these types are all different from each other and are shown in the diagram above.

It should be noted that on the Veroboard panels, the connections run e.b.c. and when the '2N' version is used, the collector and base leads will have to be crossed over to line up with the appropriate holes in the Veroboard panel.

Sockets for Integrated Gircuits

Sockets are used for all the integrated circuits and although this increases the overall cost and is not essential for the operation of the instrument, it does provide the facility for disconnecting or changing an I.C. should a fault condition exist, also it does allow various parts of the instrument to be tested without incurring the expense of buying all the I.C's at once, 16-way sockets are required for I.C.8, 10, 12 and 14 and 14-way sockets for all the others.

The instrument is built using a simple aluminium chassis, panel and cover type of construction, as shown in Fig. 14. The finished chassis size is $8 \times 7 \times 3_8$ in deep and has a front and rear panel, each size 8×3 in, fastened to the front and rear edges with 6BA screws and nuts.

(To be continued)

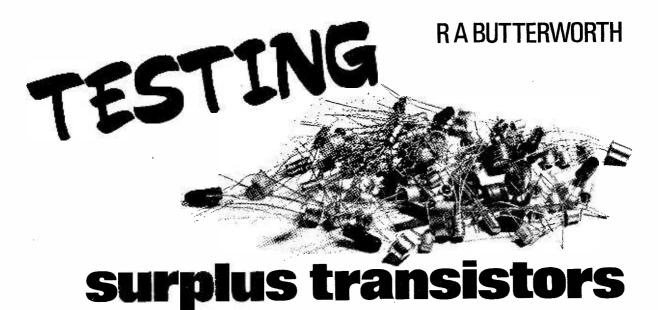
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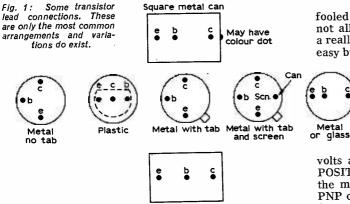


THE technical press offer so many transistor projects that many of us have bought and acquired a fair stock of transistors gathered from bargain packs or labouriously removed them from surplus computer boards. We then go through them only to find that many are un-marked, indecipherable, or have some queer markings not contained in any lists. The manufacturers are reluctant to supply data because they have been made to "customer requirements". A few hours spent in sorting these into NPN or PNP and grading them into four categories will save a lot of time, patience and money on the next project development. The methods of grading are only approximate and there are only four grades:—

- 1. RF types.
- 2. Low gain types.
- 3. High gain types.
- 4. All others which do not pass the tests described.

PRELIMINARY TESTING

First do a thorough visual examination and sort the transistors into known and unknown. The knowns, however, should be put through the tests since they may be blown or below spec. If your collection contains transistors removed from computer boards, the leads will be short. Fortunately when these boards are assembled, small plastic



Square metal can

spacers are fitted, in most cases to comply with manufacturer's recommendations as to bending and heat dissipation. The makers also usually bend over about 1 _{gin} on the solder side with the result that the lead lengths are in a safe condition. Any that are too short should be thrown away. Figure 1 shows the bases of seven categories into which these recovered units fall. Some may have a coloured dot, usually red or white, indicating the collector lead, also some are internally connected collector-to-can. Do not be

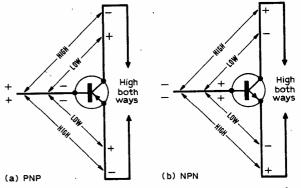


Fig. 2 : Resistance readings found on a good transistor—see text for details.

fooled however by coloured sleeving on the leads; not all manufacturers use the same code. Of course a really good transistor checker can make things very easy but this will cost a good deal of money.

May have

olour dot

IDENTIFYING THE LEADS

The first checks, sorting the PNP and NPN types and the leads are done with a multimeter on its low ohms range. A word of warning first about that multimeter. Most meters on the ohms range give positive

volts at the NEGATIVE LEAD and negative at the POSITIVE lead, so check first. Figure 2 illustrates the method of finding out whether the transistor is PNP or NPN. Never be scared of reversing the leads from the meter; with about 1.5V, which is usually available from the meter on low ohms ranges, you won't do any harm. Should you get a low resistance reading both ways between base and either of the other connections you can be fairly certain that there is an internal short. If you get a low reading between the base and one lead and a high at the other it is either bad or you are not on the base. It is probable that you will meet with marginal readings at times, so make tests as follows with reference to Fig. 3.

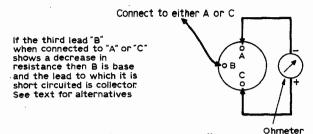
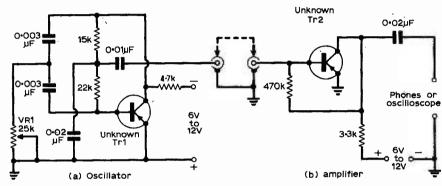
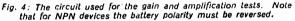


Fig. 3: Tests for identifying the base connections.





Draw a base diagram of the base of the transistor and mark them 'A', 'B', 'C'; you are now faced with three permutations, of pairs 'AB', 'AC' 'BC'. Now take any pair and connect to the ohmmeter on low ohms range, then connect the other lead to each lead in turn: if the resistance falls the unpaired lead is the base lead and the lead to which the base was shorted is the collector. Check to see if this lead is connected to the metal can, this could confirm that it is the collector. If you do not get these results then check each of the three pairs using the same procedure. If the base cannot be identified for certain and there is high resistance both ways it is most probable that the third lead is base. To check, assume this is base and check with the meter whether it is a PNP or NPN in Fig. 1. You should then have established the base connection; the emitter and collector connections will be identified by reversing the leads in the grading test.

TESTING FOR GAIN ·

To test for gain two simple circuits are required as shown in Fig. 4. Good quality transistor sockets are essential since we are most likely dealing with short leaded transistors and soldering in and out will become tedious. The assembly of the the circuits is an individual choice, though the author used 'S-Dec' and found it ideal. When the test rig is assembled, put in two good known transistors, say an OC44 or OC45 for Tr1 and an OC70 or OC72 for Tr2, connect up the battery and headphones. If you can use a 'scope, all the better. Now adjust VR1 until a tone (or trace) is observed, then start to substitute the unknowns for Tr1. If a tone is present the gain will be more than 30, if no oscillation occurs then it is less than 30. At this point those transistors which have connections which are difficult to identify can be checked and identified by insertion, keeping the assumed base connection and reversing the other two unknown leads in the socket. If no oscillation occurs, keep it for the two further tests. Do not be afraid to reverse leads even at this higher voltage; if it blows then it was probably useless anyway.

AMPLIFICATION TEST

Now replace the original test transistor i.e. the OC44/45 (or one of the good ones known to oscillate) and then one by one test all those which failed the high gain test by substitution in the Tr2 position. The note heard in the headphones will tell you if it will amplify at all. If it is possible to use an oscilloscope. a little more information will be gained on this last test which is for low gain. By means of the trace

shown on the 'scope, you can classify some transistors in the large signal class because the waveform will be distorted or clipped. The assumption in the tests up to now has been that the surplus transistors are in the small signal group.

R.F. TESTS

All the transistors available should be put through the r.f. test, irrespective of whether they have passed

the previous tests. The r.f. test circuit is shown in Fig. 5. Here the reader will have to arrange for the crystal, Tr1, L1 and C1, according to what he has available, to check frequency. As a guide, with a 20MHz overtone crystal, an OC170, L1 being eight turns of 18s.w.g. on a $^{1}_{2}$ in former, tuned with a 50pF capacitor. the circuit will cover the range up to

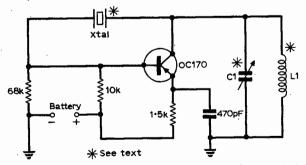


Fig. 5: The r.f. test circuit. For NPN devices the battery polarity must be reversed.

30MHz. Here a receiver or G.D.O. is necessary to detect the oscillation. Once oscillation is established it is merely a matter of substituting as before.

With your collection graded, as outlined, it is simply a matter of substitution of various types within that grade for optimum results in a particular project. The methods described are by no means comprehensive but at least you will have got a little order out of chaos and the building of many projects will be made much easier.

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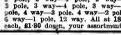
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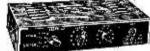
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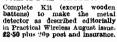
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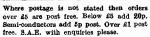
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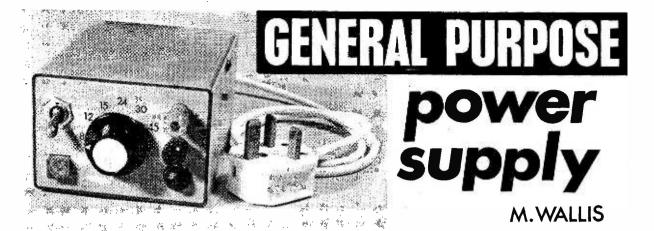




491

THE FULL-FI STEREO SIX

The amplifier



'N the "good old days" of valves, power supplies were simple. Some 90 per cent of equipment operated with an h.t. supply of about 300V and the heaters were nearly all 6.3V. A single power supply would operate the vast majority of equipment that one was experimenting with. Today the semiconductor rules the day and power supplies are very much less standardised. Checking through a recent copy of Practical Wireless showed that there were six circuits given-and five different supply voltages were specified, ranging from 6V to 30V. Batteries can cope with a number of these, especially the most common 9V supply, but even the most half-hearted experimenter needs quite a range of batteries to pursue his hobby-even for quick lashups.

One can of course build a sophisticated, stabilised power supply with variable output covering quite a large range but such circuits are expensive, are by no means simple in their construction and most of the time the sophistication is not required.

By incorporating a simple two-pole, six-way rotary switch into the circuit of a standard power supply, it is possible to build a power unit with six different outputs—all with low internal impedance and which will simulate exactly the supply that you finally build into the equipment.

A few refinements can be incorporated into such a power supply such as an indicator neon and fuse protected output which are not always necessary on the final supply which is fitted permanently but which make life simpler (and cheaper!) while experimenting.

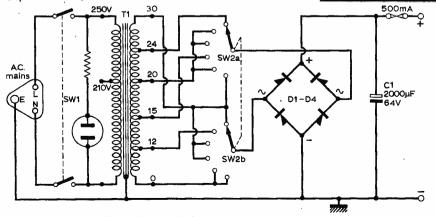


Fig. 1: The circuit of the general purpose power supply.

the circuit

The wide proliferation of supply voltages needed for semiconductor designs has brought about the introduction of mains transformers with a wide range of secondary tappings and it is simply by selecting the appropriate ones of these that various output voltages can be obtained.

The transformer specified in the components list is a good one to use but other types with multiple secondaries can be adopted.

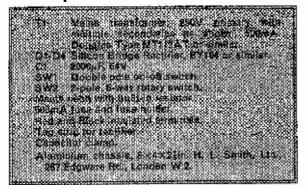
This one is fitted with 12, 15, 20, 24 and 30V tappings (these of course refer to a.c. r.m.s. volts)

/OLTS (d.c.)	VOLTS (a.c.)	Connections from which derived
4-5	3	12-15
6	4	20-24
7-5	5	15-20
9	4 5 6 8 9	24-30
12	8	12-20
13-5		15-24
15	10	20-30
18	12	0-12 or 12-24
22-5	15	0-15 or 15-30
27	18	12-30
30	20	0-20
37.5	24	0-24
45	30	0-30

and a total of 13 different voltages can be selected from this as shown in the table. Those shown in the drawings, and for which the wiring is done, tend to be the higher voltages as batteries are still the best bet up to about 9V but it is a matter of personal choice which are selected.

Note that the d.c. voltages are those which should be read with no loading and are only approximate. In any case it has been found by

★ components list



experience that the secondary taps are by no means always accurate and quite large differences have been measured from the theoretical voltage but as exact voltages are rarely needed, variations are allowable.

The output voltage from the prototype, which can be seen marked on the chassis, are those measured, not those calculated, and these discrepances can only be due to winding inaccuracies as it is wired up exactly as in Fig. 1.

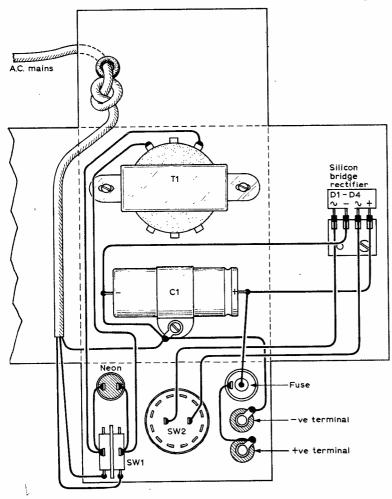


Fig. 2: The component layout and wiring details.

The current rating required from a general purpose power supply will depend on the uses to which it is put. Only power amplifiers at peaks take more than about 500mA and that was the rating chosen; it is sufficient for nearly all experimental work.

More than one company makes transformers with the voltage taps used here, but the most widely available is possibly the Douglas MT112AT. The same transformer with a 2A rating is coded MT3AT and although it will just fit into the chassis shown, it is a tight fit and a larger one is recommended.

The mains supply is connected to the correct primary (either 210V or 250V) of the transformer via a double-pole switch. A panel neon, connected

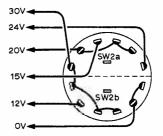


Fig. 3: The switch wiring used on the prototype.

across the primary is very useful. Today most neons are supplied with a built-in resistor but if one is not fitted, a $220k\Omega$ should be wired in series as shown in the circuit.

The earth line is shown connected to the negative line but may equally well go the positive if used in conjunction with PNP type equipment which is earthed by some other means. Whatever happens the aluminium chassis should be earthed for safety reasons.

The various connections to the voltage selector switch are taken from the appropriate tappings and the switch wipers are connected directly to the bridge rectifier a.c. contacts. Plastic encapsulated silicon bridge rectifiers are widely available at reasonable prices and come under a variety of codings. The ratings must be 50V peak, 1A. The Mullard BY164 is suitable here.

Adequate smoothing for all purposes is provided by the 2000μ F capacitor; the voltage rating of this should be well above the 45V or so applied and a 64V type was used in the prototype.

To provide a measure of protection, a 500mA fuse should be connected in the supply as shown.

construction

The prototype was built in a

small aluminium chassis size $5 \ge 4 \ge 2^{1}_{2}$ inches, fitted with a drop-in lid. These chassis are available from H. L. Smith Ltd., the address is given in the components list.

The layout of the components is shown in Fig. 2 and can be seen in the photograph.

The bridge rectifier can be mounted on a small tag strip on one long side. the transformer and the

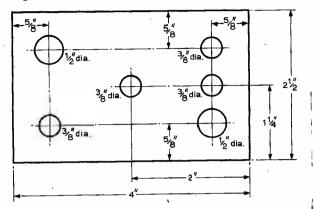
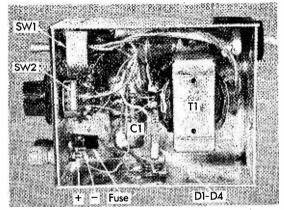


Fig. 4: The front panel drilling details.

capacitor on the base, with the remaining components on the front panel.

The switch wiring needs quite a lot of planning unless that used in the prototype is copied exactly, this is shown in Fig. 3. If other voltages, selected from the table, are required, the wiring should be planned carefully first of all. It is far easier to do the wiring on the switch leaving a few inches of wire which will finally be wired to the transformer. When all wiring is completed a thorough check

should be made to see that there is no possibility of



Internal view of the prototype showing component siting. Compare this with Fig. 2.

mains being applied to the case and that no bare wires can touch each other.

Connect a voltmeter to the terminal sockets and switch on, ensuring that a voltage reading is obtained in each switch position. Wrong wiring on a switch of this sort can lead to a direct short of the secondary and—unless noted very quickly—unpleasant things can happen to both the transformer and the switch.

There may be a very small amount of sparking while switching but this does no harm. The prototype has been in use continuously for several months with no ill effects. There is just one point that must be made. The charge stored in a smoothing capacitor of this size lasts for a long while and if the power supply has been on a high range the capacitor should be shorted out (with the mains switch off) before being applied to a circuit where a high voltage could cause damage. This should be done through a resistor, not directly, as the fuse may blow otherwise.

If, due to the nature of the experimenting, a considerable amount of switching is envisaged, a resistor can be wired permanently between the output terminals so that a small current is drawn at all times and the capacitor will then settle down very quickly to the applied voltage. A $4.7k\Omega$ resistor with a ${}^{1}_{2}W$ rating will serve this purpose.

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The Practical Wireless Crossword Competition No. 2, published in the June 1971 issue, has now closed and the prizes of £10 vouchers, exchangeable with any advertiser in Practical Wireless have been awarded to:

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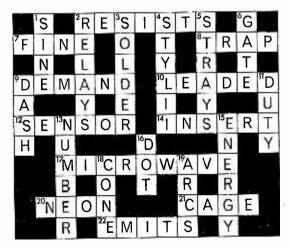
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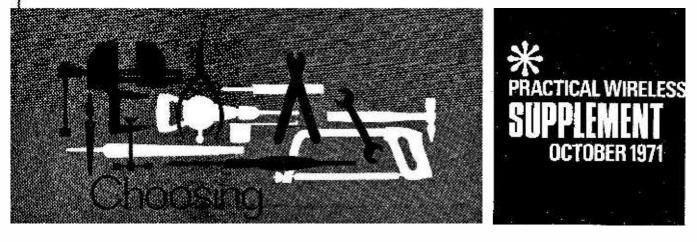
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SMALL WORKSHOP TOOLS S. YEATMAN

SINCE YOU ARE reading Practical Wireless the chances are that you have already tried your hand at constructing one of our projects. If you are already well fitted out with tools or have access to workshop facilities the bending up of a piece of aluminium to make a chassis or the punching and drilling of various holes would not have presented you with much of a problem. But if you are a newcomer to home electronics you have probably employed 'knife and fork' techniques, utilising kitchen knives and similar tools!

This really is doing things the hard way and the effort involved can put a considerable damper on anyone's enthusiasm for home construction, apart from any ensuing domestic problems when the lady of the kitchen discovers that her best kitchen knife now looks like a hacksaw blade!

If you are really taken with the idea of constructing electronic gear then be prepared to start off properly by buying a few good essential tools designed for the job in hand. Forget Dad's woodworking screwdriver with its $\frac{1}{2}''$ wide blade filed so thin that it seems to fit any screw head from 8BA upwards! Very convenient, of course, but what a mess it makes of screw heads! Dad's pliers, too, are very handy—they 'fit' any size of nut you like to mention but by the time they have slipped a couple of times that nut looks pretty sorry for itself.

When you find out what fun these electronic projects can be you will begin to take pride in the finished job and Mum's knives and Dad's tools just won't be good enough. So, start off on the right foot by getting the proper tools and remember that, like most 'bargains', cheap tools can turn out to be expensive in the long run. Spend a little more on tools from manufacturers of repute and scorn imported tools made of inferior materials.

The following review lists small workshop tools more or less in the order of necessity bearing in mind the requirements of the home constructor. The soldering iron takes precedence since even the simplest bought kit of parts needs the odd soldered joint or two.

In a review of this kind only general reference can be made to the great variety and range of tools that are available today. Specific recommendations are impossible without some knowledge of the amount of money it is intended to spend and the precise nature of the work to be undertaken in the workshop.

Workshops tend to 'just grow' from a few simple tools. As the limitations of these tools become apparent so more tools are bought to fill the gaps, thus it can be seen that a large initial outlay is not necessary. Spreading the cost over a period allows emphasis to be made here on the requirement to buy quality, a policy which, if followed, will not be regretted.



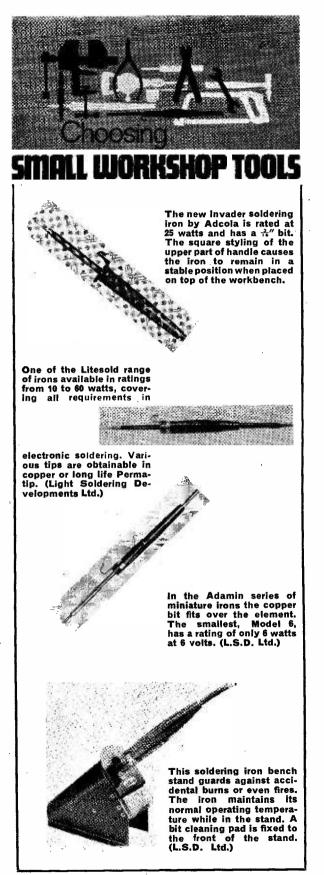
SOLDERING IRONS. There is a wide variety of electric soldering irons available today, the main problem being the selection of one of a suitable wattage for the home constructor of radio gear. For the average Practical Wireless project a 25 watt iron should prove satisfactory but if a lot of work is likely to be done on printed circuit boards a 15 watt iron would be more appropriate. These circuit boards often contain many small components fitted quite close together and thus vulnerable to the heat of a soldering iron. Excessive heat can also damage the copper strips of the board itself.

With these irons goes a wide range of copper bits of differing diameters, some straight and some angled. Initially a straight bit $\frac{1}{6}$ " or $\frac{3}{16}$ " in diameter can be chosen for general purpose work with an angled bit for the awkward spots. At a little extra cost special bits are available which contain additives to the copper designed to prolong the life of the bit.

For home use a soldering iron operating directly from the 230 volt mains supply is usual but, for anyone who might feel a little happier with a lower working voltage, irons are available to work from as low as 6 volts via a mains isolating transformer.

When wiring an electric iron to its mains plug ensure that the instructions supplied with the iron are strictly adhered to so that the iron is properly earthed. This is a very important point for the builder of electronic equipment when wiring in transistors and other solid-state devices. Irreparable damage can be caused to these devices by stray earth currents if the soldering iron is not properly earthed.

Advancing rapidly in popularity are the soldering guns or 'heat guns', as they are sometimes known. The copper bit is, in fact, part of a low voltage secondary winding of the built-in transformer. Exaggerated claims are sometimes made concerning the short heat-up time for these guns,



typically 10 seconds, but the better known makes can be relied upon in this respect.

The trigger style on-off switch on these guns often has two positions, for fast heat-up and for running heat, thus giving a dual-heat facility. Sometimes other types of bit are available which can be used to cut sheet plastics or to weld plastic sheet and so on.

Solder today, as used on electronic work, invariably has several cores of flux to ensure better distribution of the flux around the joint and consequently a better joint. The solder contains tin and lead in varying proportions which affect the temperature at which the solder will melt. The usual ratio of tin to lead is 60/40 for radio work and the diameter of the solder can be between 18 and 22 s.w.g. depending upon the fineness of the work being done.

While short lengths of solder can be bought for a few pence the $\frac{1}{2}$ or 11b. reel will be found to be more economical for the home constructor. If however he wishes to experiment with various solders he will find many easy-to-use dispensers and handy-packs available with different types of solder such as 40/60 or 60/40 tin-lead alloys as well as solders containing a small amount of copper which is calculated to reduce the amount of wear on the copper soldering bit.

Although soldering irons are usually fitted with a hook of some kind so that they can be hooked onto a convenient point it is a good policy to provide some kind of metal stand for use on top of the bench. Such a stand can be easily made out of an odd piece of aluminium or tinplate.

Soldering iron manufacturers naturally produce their own stands or 'protective shields' for their irons and these are designed to ensure that the iron does not overheat when the iron is not being used but is still switched on. For convenience a cleaning pad for the bit may be incorporated and possibly a holder for a reel of solder.

A 'must' in soldering is a sharp pick or long needle for clearing out holes in soldering tags and removing surplus solder. Commercial versions are usually made of aluminium, to prevent the solder sticking to the needle, and with a plastic handle or covering. One end may be pointed and the other end split to enable wires to be unravelled from tags etc., while the solder is molten.

Another de-soldering aid available is in the form of copper wire braid. When the end of the braid is touched on a hot soldered joint surplus solder travels up the braid by capillary action. The end of the braid is snipped off ready for the next joint.



PLIERS AND CUTTERS. Again a very large field of choice and again the rejoinder to pay a bit extra to get quality. Two pairs of pliers will suffice for most constructional work, a long-nosed pair with tapered jaws coming to a reasonable fine point and a heavier pair with stocky jaws, maybe $\frac{1}{2}''$ wilde, and probably incorporating a pair of sidecutters.

The long-nosed pliers will always be in the hand of the electronics constructor, for twisting wires round tags and forming the wire ends of components before soldering them into position. Don't forget that a wiring joint should always be mechanically sound before it is finally soldered, although the true experimenter will not go along with this principle because he will want to use the same components time and again!

Variations on the pliers theme include those with angled

jaws of various lengths, round jaws and flat narrow jaws rather than pointed ones. When buying pliers see that the opening and closing action is not too loose or the jaw tips will not close together properly and it will be difficult to get hold of fine wires when wiring up.

Wirecutters will be used as often as the pliers so, again, only the best must be chosen. They should be about the same size as the long-nosed pliers with the cutting edges at an angle to the body. Other types have the cutting edges across the top or parallel with the body.

When buying sidecutters hold them up to the light and see that the cutting edges come together cleanly and evenly throughout their length. Try this test with a cheap imported pair of cutters and see the difference! The previous remarks concerning the action of pliers applies just as much to sidecutters. A loose action and the cutting action will be greatly impaired. On the other hand, there is nothing more annoying than a pair of cutters or pliers that need two hands to open them!

WIRESTRIPPERS. Different people have different ways of stripping the insulation from wire, varying from the use of the front teeth to an old kitchen knife! If the wire is nicked, and it often is, it will eventually break if it is subject to vibration or flexing. The ordinary sidecutters are frequently pressed into service as wirestrippers mainly because they will have just been used to cut off a piece of wire and are still in the hand! However there are plenty of wirestrippers on the market designed for the job so there Is no excuse for improvisation.

Wirestrippers usually have some means of adjustment to cater for different size wires. It is important not to try to strip wires of a larger gauge than that for which the wirestripper has been set, or the wire will be nicked. One of the latest style wirestrippers on the market has a four position cam that can be adjusted immediately to another gauge of wire, while still held in the hand. Incidentally, this model also has wire cutters which have ground edges, a great improvement.

A feature on some types of wirestrippers, which the author considers of great importance, is the provision of a spring that keeps the cutters normally open, the cutting action taking place against the spring.

SCREWDRIVERS. Strictly speaking, every different sized screw needs its own screwdriver, if the blade is to fit the screw slot properly, but not many of us are likely to go to that extreme. The width of the blade should be about the diameter of the screw head or wider, rather than narrower. For electronic construction work it is best to buy a set of screwdrivers with blade widths from about $\frac{1}{4}^{"}$ up to $\frac{1}{4}^{"}$ or $\frac{5}{16}^{"}$. Three screwdrivers should be sufficient. A useful addition is one for the Philips style screws, that is the screws with crossed slots.

Screwdriver sets are available consisting of several blades that can be fitted to a common handle. Although these may be cheaper than individual screwdrivers they are less convenient in practice.

The screwdriver blade itself should be about 6'' or so long depending upon the blade width. One very useful size of blade is one about $\frac{3}{32''}$ wide for use with knob setscrews. The blade should be as long as possible so as to reach a knob in spite of the clutter of other knobs, dials etc. on a panel.

Don't take any notice of the markings on some screwdrivers that seem to indicate that one can use them with safety on 10000 volt circuits! Anyone having that sort of voltage around should know the precautions he should take. They would not include the use of a screwdriver on 10000 volts!

Variations on the basic screwdriver include the ratchet

A most useful de-soldering aid is a reel of specially treated copper braid which will completely remove unwanted solder from a joint by capillary action.

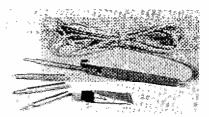




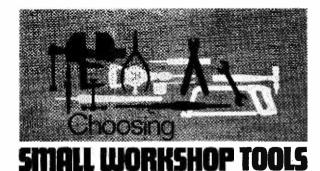
The Adcola protective shield is intended for wall or bench mounting and has slots for a spindle carrying a reel of solder.

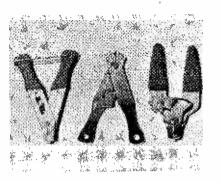


This de-soldering tool (L.S.D. Ltd.) could be invaluable when stripping components from surplus equipment. It is self-contained and simple to operate and may be used as a conventional soldering iron.

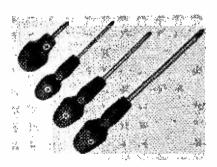


The Adamin miniature iron, Model 15, can be obtained complete with a set of four bits of differing size. The tube of special lubricant is used to prevent the bits from sticking to the heating element.





Three of the Bib range of wirecutter, strippers by Multicore Solders Ltd. The latest, centre, is the Model 3A with an instantly adjustable cam enabling the strippers to deal with most wire sizes. The cutters have ground cutting edges. The ends of the arms are, in fact, flat spanners for 4 and 6BA nuts, a most unusual feature.



A set of Stanley 'Pozidrive' screwdrivers enabling Philips type screws to be reached in the most awkward corners.



The Stanley range of 'Spee-D-Grip' screwdrivers incorporating a spring clip which holds screws in position for starting in inaccessible places. types which remove the necessity of continually lifting the screwdriver from a screwhead thus, incidentally, reducing the risk of the blade slipping and doing damage. Of course, if one used a screwdriver properly, the blade should never leave the slot, but this seldom happens!

A most valuable addition to a set of screwdrivers is one having a spring attachment at the tip of the blade to firmly hold a screw while it is being inserted into place. Its value in awkward spots is inestimable. Such an attachment can sometimes be bought separately for use on almost any screwdriver.

SPANNERS. There is always a temptation to use a pair of pliers on any nut that needs tightening, usually because the pliers are ready to hand. It should be remembered that the opposite faces of a nut are parallel to each other where-as the jaws of a pair of pliers are only parallel when they are closed. That's why the pliers always slip off the nut and cause damage.

First of all, get a set of box spanners covering 0 to 8BA either single or double ended. The single ended ones with a wooden or plastic handle are recommended. Just two, for 4 and 6BA, are worth their weight in gold in a workshop. Having long stems they are invaluable for placing nuts into odd corners where fingers or pliers cannot reach.

It is always good practice, wherever possible, to tighten a nut on to a screw rather than use a screwdriver on the screwhead. This is particularly important where the screwhead is visible such as on the front panel of a piece of equipment where a damaged slot on a screw would be an eyesore.

Another requirement in the spanner line is a set of flat spanners, again covering the BA sizes, possibly 0 to 8BA. These can very well be double ended ones since the length of the spanners is not of much importance.

Some sets are single ended but bolted together at one end so that one can be selected and the others swung out of the way, rather like a set of feeler gauges.



DRILLS. In the past a hand drill would have been a first acquisition in a workshop, followed by an electric drill when the money was available. Today the cost of electric drills is such as to justify buying one from the outset.

The simple electric drill will have a single speed and accommodate drills up to t' in diameter, ('twist drill' is really the proper name for the bit that does the cutting). This model will perform just about all the routine drilling jobs in a workship where electronic projects predominate. A chuck adaptor to take drills up to t' in diameter is highly desirable, enabling the usual holes for mounting potentiometers etc. to be drilled without having to resort to files or reamers.

If the drill is likely to be used for other duties around the house then the two speed drill with a chuck taking drills up to $\frac{1}{2}$ " in diameter can be considered. A popular accessory for the electric drill is the speed controller which is an electronic device enabling the speed to be varied continuously from a few r.p.m. to maximum speed. Such controllers are available in kit form or ready assembled.

If a lot of drilling is anticipated, particularly on chassis work, a drill stand will prove a boon. Most electric drill manufacturers have such stands but their design is such that they will accommodate most makes of drill. Such stands ensure that holes are drilled vertically, which may not be very important on sheet metal work but could be critical with a deep hole. Holes will be more accurate both in size and positioning when drilled on a stand rather than with a hand held electric drill.

When using drill adaptors enabling, say, a $\frac{3}{2}''$ drill to be used in a chuck of nominal $\frac{1}{4}''$ capacity, remember not to overload the motor by exerting excessive pressure on the drill especially in hard materials.

Hand operated drills suitable for the small workshop range from the simplest single speed ones with a $\frac{1}{4''}$ capacity to those taking drills up to $\frac{1}{4''}$ and having an integral gearbox providing either a low speed or a high speed at the chuck. A little extra spent on a good hand drill will be a worthwhile investment. In particular the gears will run more smoothly and require less energy and the chuck will be of a better quality.

Now, what about the twist drills themselves? These are bought singly, as required, or, if starting more or less from scratch, then in a set complete with a stand for use on top of the bench. They may be had in sizes from about $\frac{1}{32''}$ to $\frac{1}{4''}$ in diameter or in a numbered range from No. 80 (0-0135'') to No. 1 (0-228''). Letter drills carry the range on upwards from 'A' (0-234'') to 'Z' (0-413'').

For many years the author has had just four drills always to hand, No's 12, 24, 32 and 43. These will cover the drilling of holes for tapping or clearing for 2, 4 and 6BA screws.

As far as the quality of twist drills is concerned, it is necessary to take a little more care over their selection than with other tools since there are a lot of drills on the market that seem to be made of little more than soft iron! So once again beware of cheap imitations and if the pocket will stand it, buy a small set of genuine high speed drills. These will last for ever and seldom need sharpening if not abused.

PUNCHES. Whenever it is required to drill a hole in a chassis, or anywhere else for that matter, it is essential to use a centre punch to mark the spot where the hole is required. Otherwise it is almost certain that the drill will wander off the spot, scratching or damaging the surface being drilled. This particularly applies to hard surfaces and when using a hand held electric drill especially if care is not taken to keep the drill vertical.

There's not much to say about centre punches. A good one will be of tempered steel with a sharp round point, usually machine ground. Spring loaded punches are available which obviate the need for a hammer, so these punches are very convenient and the extra cost well justified.

Chassis punches are a necessity although the number bought for a small workshop will probably depend upon the funds available. These punches will cut holes ranging from $\frac{3}{4}$ " to 3" in diameter but it is recommended that at least two be obtained, $\frac{3}{4}$ " diameter for B7G valveholders and $\frac{3}{4}$ " diameter for B9A valveholders. These punches require a suitable Allen key to operate them and their cutting limit is about 16 s.w.g. mild steel. They will cut through the usual 16 s.w.g. aluminium sheet like the proverbial knife through butter, thus saving a great deal of time and energy.

A useful tip ... the resulting hole will have a burr on one side so place the cutting part of the punch on the side of the panel or chassis that might be of some importance from the point of view of appearance.

Other chassis cutters are available that will cut square or rectangular shaped holes but the infrequent need for such cutters does not justify their purchase.

The predecessor of the chassis cutter was the tank cutter. This had an adjustable cutting bar for different diameters of hole, but unless used in a lathe or drill press,



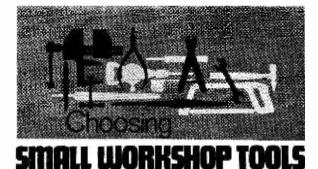
This bench clamp enables a portable electric drill to be used with grinding, polishing or other attachments (Wolf). The drill is a Wolf Sapphire 73.

No market

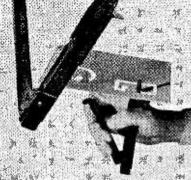
For precision drilling a bench drill stand is essential. The spring loaded press magnifies the hand pressure some ten times. (Wolf)



The old-fashioned hand drill now has a streamlined appearance. In this Stanley drill the gears are completely enclosed protecting them from dirt, swarf etc. and obviating nipped fingers!







Requiring only a $\frac{1}{4}$ " hole to begin operations this Adel nibbling tool will cut a hole of any shape or size in sheet steel up to 18 gauge steel or aluminium up to 16 gauge. (West Hyde Developments Ltd.) The punch and die action cuts without distorting the metal.



This precision made Mole 'Supercut' tool has a powerful action capable of cutting long straight lengths, curves or angles in sheet materials such as steel, aluminium, plastic laminates etc. A replacement blade can be fitted very easily.



The standard Mole wrench has many uses in the small workshop. In spite of its immense gripping power it can be opened at a touch on the release lever. the resulting hole could be a little rough in appearance. A slow cutting speed was essential and for this a carpenter's brace was ideal.

Other hand tools for cutting holes in sheet metal include the family of "nibblers'. A small hole is first drilled in the sheet metal, the nibbler inserted and the action of the tool nibbles the material away to form a hole of almost any shape desired. The hole can be finished off with the usual files.

Shears, or tinsnips, will be needed for cutting sheet metal and a 10" pair with straight jaws should suffice for general work. The will make a straight cut in sheet metal for a limited distance although the 'goosebill' type will usually prove better for this work.

TAPS. Now and again the necessity will arise to tap a thread in a hole and for this two taps for each size thread are required, together with a tap wrench. The first tap is tapered, to start the thread, and the second tap is full size. Sometimes a third tap of intermediate taper is used but this is not really required for small threads.

The tap sizes most likely to be required are the usual 2, 4 and 6BA. Taps can be broken off relatively easily and for this reason a proper tap wrench is strongly recommended.

The simplest and probably the best tap wrench is the bar type with a screw grip on to the tap. The chuck model is more expensive but it does enable the tap to reach positions that would be inaccessible with the bar type wrench.



FILES. For finishing off the edges of metalwork and suchlike a 10" flat file will be quite adequate. That known as 'second cut' will prove best being neither too coarse nor too fine for work on soft metals. A similar sized half round file is very useful for finishing off large holes such as those required for meters. While smaller holes, for valveholders etc., can be cut with chassis punches the cost of similar punches for large meter holes tends to be prohibitive, especially as they are used relatively infrequently. Such large holes can be made by drilling a circle of small holes, then knocking out the disc of metal and finishing the work with a half round file.

A $\frac{1}{4}''$ round file completes the essential complement of files. This will find most use in enlarging small holes in the absence of a reamer or for 'drifting' holes that are not quite where they ought to be! Perhaps the spot was not properly spotted with a centre punch before drilling!

With the advent into common use of slide switches on electronic equipment the need has arisen for a couple of small needle files, one round and the other square, for finishing off the rectangular holes required in panels for these switches. If a set of needle files can be managed then so much the better as these sets contain files of circular, triangular and flat sections as well as those already recommended.

When buying files get suitable wooden handles for them at the same time. Don't try to use a file without a handle or there will be a grave risk of the tang of the file entering the wrist if the file should jam or stick in the work.

HAMMERS. As far as hammers are concerned the small workshop should have one with a plastic or rubber head for bending sheet metal, without causing undue marking of the metal. A second hammer should be a light one of the 'ball-pein' variety for general work around the place. A good make is advisable to ensure that the hammerhead is properly attached to the shaft as a hammerhead in orbit around the workshop is not funny!

REAMERS. During the construction of a piece of electronic equipment one always seems to need to drill a hole that is between the sizes of the drills available. In such a case a tapered reamer can be invaluable, as an alternative to filing out the hole with a round file, a process that does not endear one to one's family, if they are within earshot!

A pair of reamers can deal with holes between $\frac{1}{4}''$ and $\frac{1}{4}''$ in diameter producing a clean circular hole with little or no burr. They are usually intended for use in a hand brace but can on occasion be held in the hand.

VICES. If a bench or table is permanently available for constructional work then a vice of some sort will prove to be a great boon especially when fabricating metal work. While smaller sizes are obtainable, an engineer's vice with 4" wide jaws should be the smallest size to be considered. Such a vice is extremely rugged and will take a lot of punishment, including using it as an anvil!

Since the vice jaws are made of hardened steel, and serrated at that, it is essential to make up a pair of soft jaws to protect work held in the vice. These jaws can be made from pieces of aluminium or other soft metal. Other vices have clip-on jaws which are plain, serrated or slotted to hold tubing or round material more securely. Another model of vice has a separate mounting plate that can be attached to almost any flat surface by means of a suction pad.

A carpenter's vice with 6" or 8" jaws is even better for sheet metal work since the jaws can be opened to a far greater extent than those of an engineer's vice. However the space needed to permanently mount a carpenter's vice, especially below and behind the vice, is considerably more than that needed for an engineer's vice. However, there are some portable vices on the market that seem to combine the best points of both types.

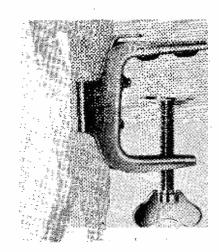
Although one may possess a bench vice there is always a need for a portable clamping or holding device and this is where the wrench comes in very handy.

The modern version of the wrench has an easily operated lever that allows the wrench to be locked on to the work and just as easily released. A clamp to hold the wrench is also available, which can be fitted to a table or bench top with a single clamping screw, thus providing a fixed vice if necessary. The jaw opening of the wrench can be adjusted over a wide range by a single screw fitting.

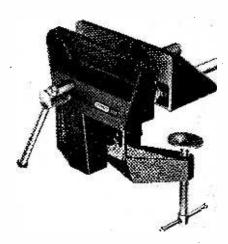
RULES. For marking out on sheet metal a rule and a square (trysquare) will be required, if an accurate job is to be done. A 12" rule is good enough and although it would seem that choosing such a simple tool would be no problem, the choice is, in fact, very wide indeed, ranging from rather inaccurate wooden rules, with printed graduations, to engraved steel ones.

Although a 6" square can be bought separately it really pays to buy a combination square since this incorporates a rule and square, usually of reasonable quality.

HACKSAWS. The hacksaw will come in for quite a bit of use in the small workshop, mainly for cutting sheet metal to size before forming into a chassis. Some of the cheaper hacksaw frames will be found to twist when the blade is tightened up, so a good quality frame will be a sound buy. The adjusting nuts, too, can be difficult to turn owing to poorly formed threads.



The versatility of the Mole wrench can be greatly increased by the use of this table or bench clamp, converting the wrench into a portable vice.



This Stanley aluminium portable vice has 5" jaws and can be clamped to any convenient surface. It is eminently suitable for bending sheet metal into chassis etc.



The Mini-Mole is just 5" long and very useful for getting into those tight corners, inaccessible to its 'big brother'. The curved jaws provide maximum contact with round or irregular shapes.



A frame to take 10" or 12" blades is big enough and the blades themselves can be of the low tungsten flexible type. These will deal with general cutting of mild steel, brass and aluminium etc. and absorb a lot of misuse without breaking. For maximum performance high speed flexible blades are to be preferred although they are a little more expensive.

For maximum efficiency the size of the teeth on the blade must be related to the material being cut. For soft metals, such as aluminium, a coarse blade is desirable to prevent the teeth from clogging, but on harder metals such as brass or on thin sheet metal a finer toothed blade is required. The number of teeth per inch of blade normally runs from 14 to 32 for hand held saws.

A very useful addition to the standard hacksaw is the 'junior' or bow-frame saw in which the blade is held under tension by the bow. The standard blade is 6" long with fine teeth. This saw is very handy for the smaller, finer jobs where the standard hacksaw would prove clumsy. In the electronic workshop the bow saw will be used a great deal for shortening the spindles of potentiometers, capacitors etc. Remember to hold the unwanted end of the spindle in the vice while cutting, rather than the component itself.

TRIMMING TOOLS. Although perhaps it is wandering into the realms of servicing it is not inappropriate in this review to mention trimming tools for electronic equipment. After all, having built a lovely radio set in our nice new workshop, it won't be of much use unless the tools to align it are to hand.

These trimming tools can be bought individually, as needed, but preferably they ought to be obtained as a kit. This will prevent the use of ordinary screwdrivers for adjusting iron cores which can only cause damage in the long run.

A kit may consist of one basic tool holder with several types of trimming tool that can be fitted into the holder or, in the more expensive kits, a complete trimmer for each purpose, such as adjusting iron cores, capacitors and special types of capacitors, such as the Philips pre-set beehive capacitors, which are met with quite frequently.

In conclusion it should be pointed out that for every basic tool mentioned in this review there are a dozen or more variations on the market. Like women, there are short ones, tall ones, thin ones and fat ones! Each one to his own fancy! The average D.I.Y. or tool shop is sated with gadgets that promise to relieve us of the tedium and work required to do a particular job. Few succeed, and those that do probably take twice as long to set up to do the job as one would take using basic tools alone.

It's surprising how often one resorts to basic methods and tools in the workshop. If funds are limited then put them to good use by buying only the best in tools from reliable manufacturers. If their names happen to be a household words they would not have reached those dizzy heights by selling rubbish.







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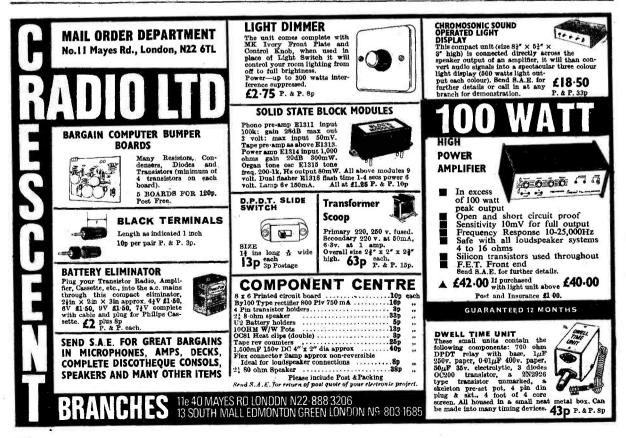
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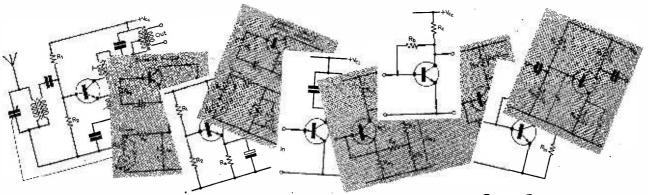
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All this



TRANSISTOR CIRCUITRY for beginners'

SPECIAL SERIES-READ IN CONJUCTION WITH OUR FREE WALL CHART!

WHEN the idea for this series was first broached, everything seemed straightforward and simple. We would entitle the wallchart 'Building Bricks,' said the Editor, in an inspired moment, and the articles would take up each 'brick' and develop it so that newcomers to transistor circuitry could understand what was happening inside those inscrutable black boxes.

Which was all very fine—but it completely overlooked the fact that any beginner worth his salt is going to say 'But Why???'

No good my declaring '... an input signal to the base of Tr1 turns it on, which causes Tr2 to turn off and Tr3 and Tr4 to return to the stable mode, releasing Tr5, which now bottoms and allows ... etc.' It would sound like a paragraph from Henry!

To understand transistor action, it is imperative that we begin by understanding what makes the darned things tick, That means going back to pure physics, not too deeply, preferably without mathematics, and then progressing to the methods of construction and the reasons why there are essential parameters—rules of operation, if you like—and this is what we shall begin with.

Semiconduction

As the name implies, the semiconductor is neither pure insulator nor electrical conductor, but something in between. Its resistivity lies roughly midway between that of conductors and insulators. This resistivity, moreover, decreases with increase of temperature (the resistance of a conductor increases slightly as the temperature is raised). This is one important difference between semiconductors and conductors.

Next, perhaps more important, is the way in which current flows in a semiconductor. There can be two processes; either a flow of negative electrons, as in the case of a conductor, or a flow in the opposite direction of positive 'holes'. The concept of hole flow is so vital to the understanding of transistor operation that we must begin with a simple discussion of atomic structure in order to get a grasp of it.

Atomic Structure

An atom consists of a central nucleus containing positively charged protons with a number of negatively charged electrons orbiting around it. The negative charge carried by an electron is equal to, and exactly opposite to, the positive charge of a proton. In a neutral atom, the numbers of each are equal.

Different elements have atoms containing different numbers of electrons, orbiting at different distances from the nucleus. The electrons close to the nucleus are tightly bound to the nucleus, but those circling in the outer orbits are more loosely bound, and are easier to dislodge from the parent atom. The outer ring is called the **valence ring**, and the valence electrons are those with which we are mainly concerned in this work. The looser they are bound to the atom—in any material—the better a conductor a piece of that material will make. This is because loosely bound valence electrons are free to be attracted by a positive charge, such as a voltage applied across a conductor by means of a battery.

Electrons flow out of the conductor into the battery at the positive connection, but an equal number pass from the negative connection of the battery into the conductor at the negative terminal, keeping the overall total of electrons the same.

Conversely, an insulator is a material in which all electrons are tightly bound to the nucleus, so that a large amount of energy is needed to break any loose to make a current flow.

Crystal Structure

The structure of a crystal of germanium or silicon is depicted in Fig. 1. It can be seen that there are four valence electrons in the outer ring and these are balanced in the nucleus by four positively charged protons, as indicated. In the three-dimensional structure that exists in reality, every atom is equi-distant from four other atoms and each valence electron forms a pair with one from an adjacent atom.

In two-dimensional form this can be shown as in Fig. 1, with pairs of electrons forming co-valent

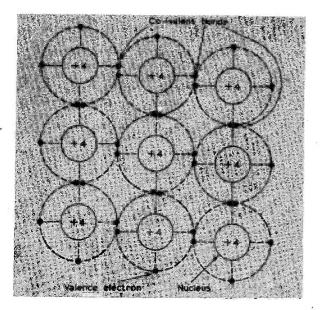


Fig. 1 : Illustrating the structure of a germanium or silicon crystal.

bonds. This is a perfect crystal, with no impurities; and no electrons are dislodged from their atoms. This is, in fact, the basic structure of a perfect insulator, at absolute zero temperature, but as temperature increases, the energy of each atom is increased and some electrons can break away. If a voltage is applied, these free electrons are attracted to the positive pole and a current flow results. The space left in the crystal structure by a dislodged electron is termed a 'hole' and this represents a net positive charge.

We can therefore think of a movement of negative electrons toward the positive voltage terminal and a virtual movement of positive 'holes' (as successive electrons move to fill the vacant spaces) toward the negative pole.

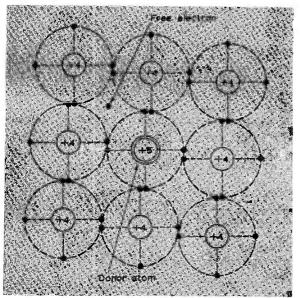


Fig. 2: Adding a 'donor' atom producing an excess free electron. N-type crystal.

N- and P-Type Semiconductor

This is a simplified explanation, but suffices to show the principle of hole movement. In a pure semiconductor there must be one hole for each free electron. Such a pure semiconductor is termed an **intrinsic material**. However, if an element of impurity is added to the intrinsic material an imbalance is obtained. If, for example, a "donor atom" having five valence electrons is added to the crystal, as in Fig. 2, there is an excess free electron. Four of the donor atom's valence electrons form co-valent bonds, in the manner previously described, but the free atom cannot do so and orbits loosely.

Some materials with a valency of five are antimony, arsenic and phosphorus. These can be added to the germanium or silicon crystal in proportions of as little as one part in a hundred million and reduce the resistivity of the germanium from 70 ohms/cm to 1 ohm/cm. Because the donor atoms in such a material contribute free electrons, which carry a negative electric charge, a material "doped" in this

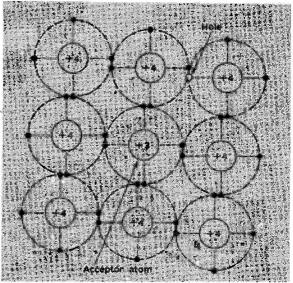


Fig. 3 : In this case, an 'acceptor' atom creates a 'hole'. P-type crystal,

way is called an **n-type** semiconductor (i.e. it possesses an excess of negative current carriers).

The opposite process takes place if indium, borium or aluminium atoms are added to the basic semiconductor material, germanium or silicon. These materials have three valence electrons, and the addition of them creates holes in the semiconductor material, as shown in Fig. 3. The resultant semiconductor material has an excess of positively charged holes and is termed **p-type**.

Minority Carriers

Minority carriers are another important factor of semiconductor physics. If a few electrons are injected into p-type material they are called minority carriers (the majority of current carriers in p-type material being, as we have seen, holes). Recombination takes place as the electrons move to fill the holes.

Similarly, holes injected into an n-type material

form minority carriers in this type of material, and small currents arise as recombination takes place. The time this takes to happen, termed the minority carrier lifetime, is an important property of a semiconductor material and the operation of most transistors depends on the movement of minority carriers.

PN Junction

A single piece of p-type or n-type semiconductor material is purely resistive: reversing the applied voltage does not alter the current flow. But if p-type and n-type regions are formed in a piece of semiconductor material, with the regions adjoining each other, a rectifying device is formed: when voltage is applied in one direction, current flows, but in the other direction little or no current flows. Such an arrangement is termed a pn junction.

In Fig. 4, the region to the right of the centre line is n-type, possessing a certain density of free electrons. To the left of the line is a region of p-type material possessing a number of positively charged holes. When the junction is first formed, it becomes an area of activity for a short period as free electrons tend to cross into the p-type region and holes move over to the n-type region.

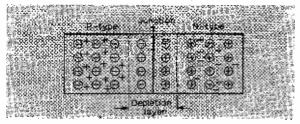


Fig. 4: Showing the charges existing in an unbiased PN junction.

This process of diffusion results in a potential building up across the junction, as the p-region gains a net negative charge with respect to the n region in the vicinity of the junction, and vice-versa, and this potential prevents further diffusion. This potential is termed the barrier potential and the region in the immediate vicinity of the junction, where there are very few free current carriers remaining after the initial diffusion, is termed the depletion layer or region, as shown in Fig. 4.

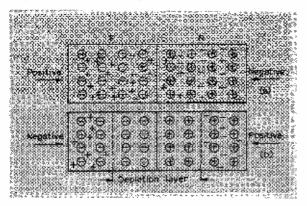


Fig. 5(a): Position of the charges with forward blasing and Fig. 5(b): with reverse blasing.

Forward and Reverse Biasing

We now have a piece of material with pronounced non-linear characteristics: when an external voltage is applied, current will flow in one direction only. If the p-type side is made positive with respect to the n-type (see Fig. 5(a), forward biasing), the effect is that the barrier potential is reduced. Majority carrier holes will diffuse across to the n-type (where they become minority carriers), and majority carrier electrons will move in to the p-type region. A relatively large current can flow across the junction in this way.

Reversing the bias, i.e. applying it as shown in Fig. 5(b), increases the barrier potential (since the external voltage adds to that across the depletion layer already) so that current flow is made very low —in fact virtually nil. However, due to the effect of temperature a number of holes and free electrons will be generated on either side of the junction—this happens even at room temperatures—and this will give the effect of a small reverse current flowing across the junction. The forward current may be quoted in milliamps but the reverse current only in microamps.

Characteristics of PN Junction

Fig. 6 shows the characteristics of a pn junction. As the forward voltage is increased from zero, the forward current increases slowly at first until the junction barrier is overcome, then rises rapidly. (In alternative terms, it can be said that the resistance of the forward biased junction is low.) In the reverse direction, the current remains low and practically

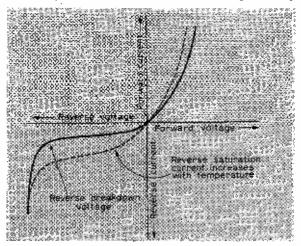
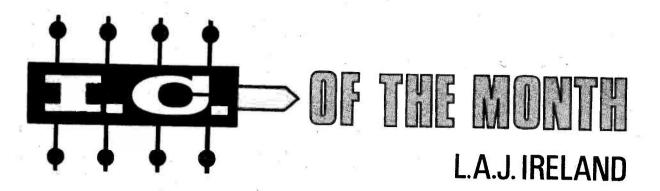


Fig. 6: The electrical characteristics of a PN junction.

constant as voltage is increased, i.e. reverse resistance is very high. Eventually, however, a point is reached where the reverse voltage accelerates the reverse current carriers to such a speed that they break down covalent bonds in the crystal structure, thereby releasing more carriers and producing an avalanche breakdown (when the junction may be destroyed). The value of voltage at which this occurs depends on the resistivity of the material.

As the reverse saturation current is due to the effect of heat, it increases with temperature, giving



Number 24

S OME months ago these notes described the Motorola MC1596 balanced mixer integrated circuit and outlined its operation as a multiplemode detector in receiver circuits. The unit discussed now goes further; full i.f. amplification facilities, at any chosen frequency up to 12MHz., are incorporated, as well as multiple-mode detection. It is the National Semiconductors (U.S.A.) type LM373, available in the U.K. from: Athena Semiconductor Marketing Co. Ltd., 140 High Street, Egham, Surrey: or Rastra Electronics Ltd., 275 King Street, Hammersmith, London W.6.

Uses

The unit is intended, not just to operate in each mode, but to make it possible to design a receiver in which intermode switching would be simple without reduction of performance. As well as the standard a.m. and f.m. modes, as used for broadcast entertainment in the m.f. and v.h.f. bands, the LM373 LM373 Multiple-mode Detector

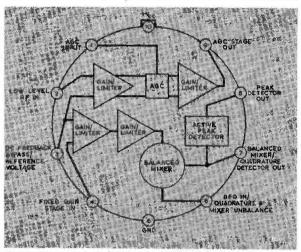
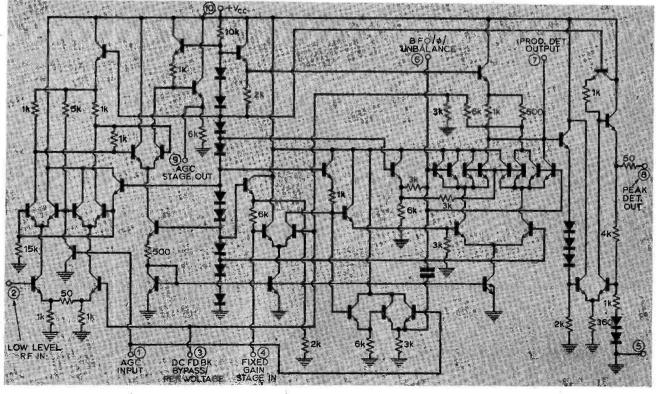


Fig. 1: The equivalent circuit of the LM373 (below) with (above) the top view of the pin connections to the various 'blocks'.



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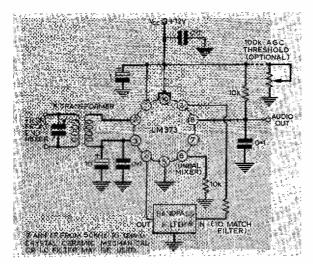


Fig. 2: External circuitry required for the LM373 when used as an AM IF strip.

can handle n.b.f.m. (narrow-band frequency modulation) increasingly used in mobile radiotelephone systems, and s.s.b. (single sideband suppressed carrier) transmissions of commercial or amateur origin.

Design

Considerable ingenuity is exercised in the design of the i.c. to reconcile the conflicting demands of each transmission system with the minimum of intermode switching in the receiver. A.M. and s.s.b. receivers must have a linear amplitude response, maintained, despite the wide range of signal strength received off the air, by an automatic gain control subcircuit. An f.m. i.f. amplifier strip, on the other hand, is essentially nonlinear, in that any input signal is amplified to a limiting level before application to the detector stage. It follows that automatic gain control[®]has no place in such a system.

To provide the facilities indicated, the LM373 incorporates four gain blocks, which can be operated in a limiting mode, as well as an automatic gain

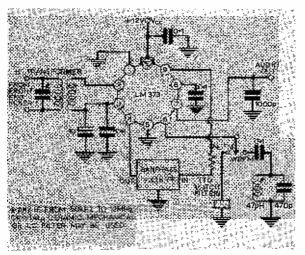


Fig. 3: The LM373 in its role of an FM IF strip suitable for narrow or wideband reception.

control system and two detectors, one a balanced mixer similar to that in the Motorola circuit already mentioned, and the other a peak detector for a.m. operation. The equivalent circuit and pin diagram given in Fig. 1 allows these sections to be identified.

Operation

Some discussion of the circuit would appear necessary. Each gain block is essentially the familiar differential amplifier, which can be operated in its linear region when the circuit is used for a.m. or s.s.b. reception, or driven to full gain as a limiter for f.m.

The automatic gain control characteristic is secured through a pair of transistors shunting the differential interstage coupling components. When no a.g.c. voltage is applied, these transistors are cut off, but as the control voltage is increased, they draw off an increasing fraction of the signal current; finally the point is reached where the signal from

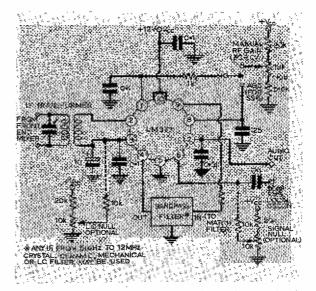


Fig. 4: The SSB/CW IF strip shown here further demonstrates the versatility of the LM373.

the first gain block is shorted out, and coupling to the second is merely through stray capacitance.

This approach secures the unusual a.g.c. range of over 70dB., enabling linear a.m. detection to be attained with input signals from 40 microvolts up to 200 millivolts. The string of reference diodes and emitter follower transistors from the d.c. +ve. line provide power supplies with decoupling to each stage of the amplifier strip, together with appropriate bias voltages to set the operating point of each element. As well as simplifying the external decoupling requirements of the circuit, these imply that each stage essentially has its own regulated supply, so minimising the effects of power line variations.

The balanced mixer stage is easily identified by the cross-coupling of its component transistors. For f.m. it operates as a quadrature detector, a system outlined when describing the MC1496, as was the product detector system used for s.s.b. In a.m. operation, this section is deliberately unbalanced by an externally applied bias, leaving only one side of the differential system connected to its load effectively.

N.b.f.m. demodulation requires a quadrature element of a higher Q factor, i.e. more rapid change of phase with frequency, and therefore a quartz crystal is substituted for the more familiar l.c. circuit. The final section is the peak detector, acting as a.m. detector and a.g.c. voltage source for linear a.m. and s.s.b. operation.

A few precations are necessary for successful operation. The a.g.c. stage output at pin 9 has a very low impedance, being taken from an emitter follower. A value of 70 ohms is quoted, so that if a high Q crystal filter is employed here a series resistor is advocated, so that the resonance peak of the component will not be damped. The manufacturer of the crystal in question will quote a recommended source impedance, allowing selection of an appropriate resistor.

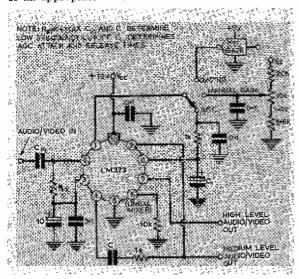


Fig. 5: In TV circuitry the LM373 can be used as an audio/video amplifier with a.g.c., manual gain control or gating.

The d.c. feedback loop at pin 3 must be carefully bypassed, since the decoupling must be effective throughout the passband of the LM373; an electrolytic or tantalum capacitor, as used for a.f. decoupling, has too high a reactance at high frequencies to be effective alone, so a small disc ceramic or silver mica is added.

For s.s.b., bypassing of r.f. at pin 7 is particularly important, as any local oscillator component reaching the peak detector would increase the a.g.c. voltage generated, and lead to an unnecessary reduction of system gain. The optional nulling circuits may be used to maximise gain should that prove necessary (an unlikely event).

Figs. 2.5 indicate the circuitry in which the LM373 may be employed, and it should be noted that the communications receiver applications for which it was designed do not exhaust the potential of this versatile unit, as is indicated by the circuit for a wideband amplifier indicated in Fig.5. Other reported applications have included use as an amplitude-modulated r.f. oscillator, an s.s.b. generator, or indeed any of the applications open to the balanced mixer circuit in which extra wideband gain is useful. So presumably more will be heard of the LM373!

TRANSISTOR CIRCUITRY FOR BEGINNERS

—continued from page 509

the effect shown by the dotted line in Fig. 6.

Junction devices of this type are known as diodes, and widely used in radio, television and other electronic applications. The diode is essentially a one-way valve, as we have seen from the forward and reverse bias characteristics.

Germanium diodes were the first types developed for commercial use, but the current-handling capacity of these devices was limited, and it was not until research provided a means of treating silicon that power rectifier semiconductors became a practical proposition. Silicon is more difficult to melt and purify than germanium, but can be doped with donor and acceptor impurities in the same way as germanium, is less sensitive to temperature, and will handle higher powers more efficiently. Common types pass currents of half-an-ampere and more with a practically negligible forward resistance. Silicon diodes have a high reverse voltage rating. (On the other hand, a germanium has greater efficiency at low voltages and at higher frequencies, and is thus retained for signal circuits.)

The pn junction forms the basis of most forms of semiconductor diode and transistor. One other form of construction is important in connection with diodes—the point contact type. In this the rectifying property is possessed by a contact between a thin metal wire and a piece of semi-conductor material. Though junction devices predominate today, point contact diodes are still much used as detectors in radio equipment. Their characteristic is similar to that of the pn junction, shown in Fig. 6, except that as the reverse current increases beyond the reverse breakdown point, reverse voltage tends to fall.

TO BE CONTINUED



Integrated circuits are beginning to appear in the new TV chassis in increasing numbers, particularly in colour receivers. The types used have been specifically developed to fulfil the signal processing requirements of TV sets but nevertheless bring with them numerous new techniques. In a new series starting this month we shall be examining these new i.c.s. to familiarise readers with the various changes in TV set design they involve.

SERVICING TV RECEIVERS

Recent articles have concentrated on fairly upto-date chassis. This month however we are going back to a widely distributed "625-line convertible" chassis, the Sobell ST195/ST282 series. This will give us an opportunity to examine the faults common in older sets.

PLUS ALL THE REGULAR FEATURES

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SERVICENTO FAULT-FINDING

PART 6

THIS MONTH G. J. KING RETURNS TO THE SERIES TO ELABORATE ON AUDIO MATTERS, DEALING WITH CONVENTIONAL VALVE AMPLIFIERS AS WELL AS TRANSISTORISED HI-FI EQUIPMENT.

N OW that colleague Hellyer has cleared the decks on alignment, superhet theory and f.m. stereo, it is my turn to have a go on audio equipment.

GRAM AUDIO STAGES

An ordinary radiogram differs little at the preaudio stages from a radio receiver, so much of the information presented earlier will also be applicable here. However, most contemporary radiograms are equipped for f.m reception, possibly with a decoder to abstract the stereo information and to direct the left and right channel signals separately to the corresponding audio channels. Thus, many radiograms and record reproducers feature a pair of audio channels and speaker systems. These are very much twin partners, the design of one following very closely the design of the other.

A mono radiogram has just the single audio channel and speaker system (though some designs employ more than one speaker unit, the mono signal being applied to them all in suitable frequency division). Whether mono or stereo, the f.m. front-end is a fairly common factor, the demodulator these days usually being the ratio detector in unbalanced or (in better quality equipment) balanced mode.

However, while the f.m. detector in a mono receiver feeds the audio channel direct (via the 50μ S deemphasis), the f.m. detector in a stereo receiver delivers the multiplex signal to the stereo decoder, the function of which is to process the sidebands of the stereo information (the difference signal) in conjunction with the mono information and the 38kHz reference signal (derived by frequency multiplication from the 19kHz pilot tone) to retrieve the left and right transmitted audio signals in the best isolation (e.g., with the least crosstalk).

De-emphasis and sometimes subchannel suppression (deleting the residual 19kHz and 38kHz multiplex components which can interfere with the audio section and add whistles to tape recordings) are engineered into the left and right audio channels, prior to the power amplifiers.

I shall be having something to say about f.m. front-ends, i.f. channels and stereo decoders later, but for now let us concentrate on the audio channels.

The simplest audio channel is designed round the triode-pentode valve, the triode receiving the detector's signal, amplifying it and then passing it to the control grid of the pentode for power amplification, Fig.1. Quite a lot of valve radiograms are equipped with this kind of audio section, as also are the inexpensive species of record reproducers. These were in fashion when stereo started, so early stereo models boast two such audio sections.

G. J. KING

The advent of stereo was concurrent with the development of transistor equipment, which was fortunate for the receiver designers, so the audio channels we now find in stereo eqipument are mostly transistorised.

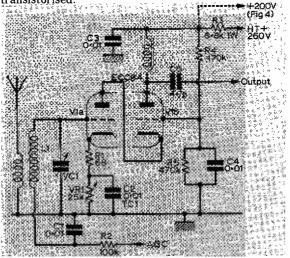


Fig. 1: A typical mono audio amplifier using a single triode-pentode valve.

Some designers favoured the use of a triode voltage amplifier (often one section of a double-triode valve) followed by a separate power pentode. This allows for greater audio power than generally possible from the pentode section of a triode-pentode valve, particularly so far as the early triode-pentode valves are concerned. Such a circuit is shown in Fig. 2, which will be seen to have much in common with that in Fig. 1. Better quality receivers have been made with a pair of triode-pentodes in each channel to provide push-pull working. In this type of circuit the two pentodes are the push-pull output valves, while the two triodes are arranged as phase splitters supplying the anti-phase drives to the pentodes. However, since few receivers like this are in the field, there is not much point in detailing the circuit, but it is worth bearing in mind.

Let us return to Fig. 1. The audio signal from pickup or radio detector is coupled by C1 to the top of the volume control VR1, the selected level of signal then being applied via C2 to the triode grid. Note that the grid is returned to chassis through R1 (10M Ω) and that the cathode of the triode is returned direct to chassis. This is a common trick to secure grid bias for the triode. Contact potential in the valve causes a minute current to flow through R1, and because of the high value of this resistor a small

potential developed across it, and since it is negative at the grid it serves as grid bias. It requires an electronic testmeter to measure this bias (since a normal meter would swamp the $10M\Omega$ with its own resistance).

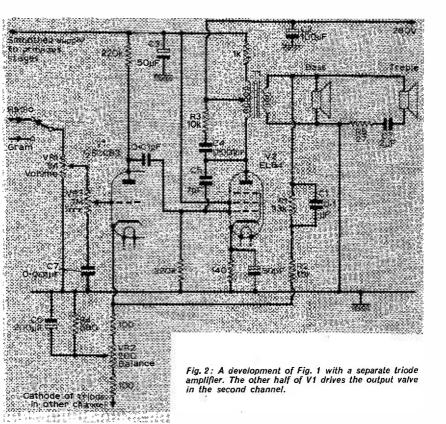
A common cause of distortion in this kind of circuit stems from a faulty triode, the fault resulting in excessive grid current and hence a potential across R1 in opposite polarity to that of the bias. The triode anode current rises, and often the valve tends to overheat.

R2 is the triode anode load, with R3 and C3 the decoupling components. The signal across R2 is coupled by C4 to g1 of the pentode, which is biased by the voltage drop across the cathode resistor R4, the grid being returned to chassis through R5. C5 is the cathode bypass capacitor (always an electrolytic in audio output stages).

The anode of the pentode is loaded by the low impedance speaker via the output transformer T1. C6 across the primary attentuates high-order harmonic distortion and makes the reproduction more palatable, as there is no negative feedback and no cancellation of second-harmonic distortion by any push-pull action.

FAULTS

Owing to the simplicity of this circuit, (Fig. 1) faults are not generally very difficult to define or locate. A cause of bad distortion is poor insulation of the coupling capacitor C4. This would permit a positive potential to exist at the control grid of the pentode with a consequent counteracting of the negative bias. It would be accompanied by excessive positive voltage at the cathode of the pentode, which could result in the failure of C5. The valve would also



run very hot, possibly with the pentode anode glowing red.

Complete failure (e.g., audio drive at the top of the volume control VR1, but no speaker output) should first lead to a valve test, followed, if necessary, by voltage checks on the appropriate electrodes. C6 is subjected to audio voltage peaks which sometimes result in it shorting. This would cut the output completely, of course, but without affecting the pentode anode voltage much, so look out for this if the valve and electrode voltages seem to be normal.

Open-circuit of T1 primary deletes the pentode anode voltage, a fault condition which is often revealed by the screen grid (g2) of the pentode glowing red hot.

Low gain or volume at maximum volume control setting, assuming normal audio drive from the detector or gram pickup, is not uncommonly caused by C5 going open-circuit or low in value. This results in negative current feedback from the audio then developed across R4. The quickest way of checking this is by shunting an electrolytic across C5.

The symptom can also be caused (possibly accompanied by distortion) by the triode anode load R2 going high. If R3 goes high or C3 becomes electrically 'leaky,' the voltage at the top of R2, and hence the triode anode voltage, will drop, which again will cause low gain and possible distortion.

Before going on to more modern transistor circuits, let us have a look at some possible fault conditions in the valve circuit in Fig. 2. The faults described for Fig. 1 circuit would also be applicable here, with one or two qualifications. First, there are two separate valves involved, the triode voltage amplifier in this case being one section of a doubletriode, the other triode serving as the voltage amplifier of the partnering stereo channel (not shown). If one channel only is in trouble one can easily compare the voltages, etc. with those at the same points in the active channel, which is a quick way of bringing the fault to light. It is also possible to substitute the same type of valves (V2, for example) between the two channels, which would tell conclusively whether the valve is responsible.

There are refinements in Fig. 2, not found in Fig. 1 circuit. For example, voltage negative feedback is applied from the secondary of the output transformer T1 to the cathode of the voltage amplifier V1, and for this to happen one side of the secondary has to be connected to chassis. The amount of feedback is determined by R1 and R2, while C1 in parallel with R1 corrects the phase of the feedback signal, ensuring that it remains substantially negative over the audio spectrum. Thus, if there is a significant rise in gain (e.g., full output at a low setting of the volume control VR1) and abnormally high distortion, then R1 and R2 should be checked for value.

C1 open-circuit could encourage instability at certain frequencies, and this is sometimes 'triggered' by transients of the music. The oscillation could be at a frequency above audio, and the only way to prove this is by connecting an oscilloscope across the secondary of the output transformer T1, preferably when the amplifier is driven by a sinewave signal from an audio generator. The oscillogram in Fig. 3 gives some idea of what the spurious signal could look like on the sinewave. The audio generator would be coupled to the grid of the triode, of course.

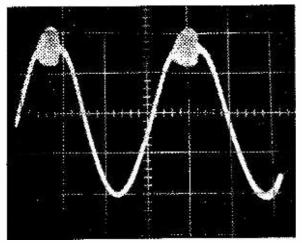


Fig. 3 : This oscillogram shows spurious oscillation triggered off by transients in the signal

If the earth and feedback connections are reversed on the transformer the feedback would be positive and the amplifier would turn into a violent oscillator! Take care, therefore, when fitting a replacement transformer! Moreover, designs like this are geared to an output transformer of specific characteristics, so if it becomes necessary to replace the transformer, and the exact type is not available, then it might be necessary to alter the feedback resistors and the phase-correcting capacitor so that the values provide the correct conditions of feedback.

An interesting feature of the circuit is the tap on T1 primary. An output transformer of this kind serves both for speaker coupling and for 'smoothing' the h.t. supply fed to the earlier stages. In fact, it acts in part rather like a smoothing choke. In the circuit, C2 is the reservoir capacitor across the power supply output and C3 is the main smoothing electrolytic.

R3 and C4 series combination across the signal part of T1 primary attenuates the higher-frequency distortion components while, with C5, providing a degree of corrective phase shift.

The triode V1 in this circuit is cathode-biased, the main cathode resistor being R4, decoupled by C6. VR2 is the stereo balance control, which feeds either side the 100-ohm resistors (one to the cathode circuit of the partnering channel). The 100-ohm resistors provide a small degree of negative current feedback since they are out of the decoupling influence of C6. The voltage developed either side of VR2 relative to the slider also contributes to the feedback. Thus it will be appreciated that by adjusting VR2 from its centre setting, the feedback on one channel will decrease, while increasing on the other channel. This, therefore, allows the gains of the two channels to be adjusted relatively for the most desirable stereo effect.

VR3 is a simple 'tone' control which works as a reactive potential-divider in conjunction with C7. With increasing frequency the reactance of C7 falls, which means that as the slider of VR3 is turned towards this end of the track the bass is lifted and the treble diminished.

It will also be seen that two speaker units are employed. That for bass is connected directly across the secondary of T1, while the treble unit receives drive via R5 and C7. R5 reduces the power, while C7 passes only the higher frequency signals. This is a very simple frequency-divider, which cannot be regarded as 'hi-fi.'

HI-FI CIRCUIT

One channel of a slightly different circuit which, possibly differing in detail, is found in some hi-fi equipment. This is the power amplifier department only, Fig. 4, where Tr1 is the input transistor, coupling to the d.c.-coupled pair Tr2/Tr3 via C1. The coupling is from Tr1 **emitter** (the collector 'earthed' to signal through C2). The stage is thus an emitterfollower, which has a relatively high base input impedance, giving the correct match from the control section.

Tr4/Tr5 constitute a complementary driver pair (complementary because one is n.p.n and the other p.n.p.). These drive the push-pull output transistors Tr6/Tr7 alternatively into conduction and cut-off on each half-cycle of signal. It will be noted that the base of Tr4 is fed from the same source (e.g., Tr3 collector) as the base of Tr5. Thus on a positive halfcycle Tr4 conducts and pushes a positive half-cycle into Tr6 base. At the same time Tr5 is driven downwards and since Tr7 base is connected this time to Tr5 collector a negative half-cycle appears at Tr7 base, pushing this transistor into cut-off. On a negative half-cycle of drive the conditions reverse, so that Tr6 is driven into cut-off and Tr7 into conduction.

The two half-cycles are reconstituted as a complete waveform in the load comprising the speaker, coupled through C3.

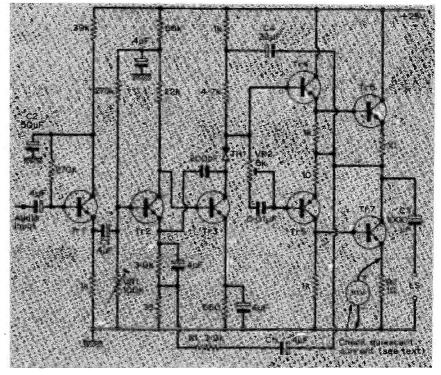
C4 is a bootstrapping capacitor which applies feedback in such a way as to increase the driver input impedance. A.C. negative voltage feedback is applied to Tr2 emitter circuit from the push-pull pair, via C5

Fig. 4: Circuit of one channel of a transistorised hi-fi stereo amplifier.

and R1. Feedback of 60dB is a common hi-fi value, which reduces the output impedance to a fraction of one ohm.

D.C. coupling exists right through to the output transistors from Tr2, so by adjusting the d.c. conditions, both the drive 'balancing' and the quiescent current of the output transistors are affected. Some circuits have no means of adjusting these parameters, so if the value of a critical component alters, so might the quiescent current and balancing.

However, in Fig. 4 VR1 adjusts the balancing and VR2 the quiescent current, which leads us neatly to hi-fi equipment fault-finding and tests. But before going on to this it is noteworthy that temperature stabilisation is often provided in or around VR2 circuit to hold the quiescent current steady with function.



increase in temperature of the transistors. The 'thermal element,' TH1 in Fig. 4 provides this

AMPLIFIER TESTS AND FAULT FINDING

For detailed testing in audio circuits, especially high quality circuits in which transistors are employed, we require an audio signal generator, audio millivoltmeter and an oscilloscope, at least. For distortion tests, we require in addition a distortion factor meter or similar instrument.

If it is assumed that the radio or pickup source signal is reaching the input of the amplifier and there is no output from the speaker, then we should check as far as possible into the amplifier with a signal tracer (e.g., pair of headphones, etc.-see Part 3, PW July 1971). In Fig. 4, for example, we could get to the base of Tr2, but possibly not any further owing to the d.c. couplings of the subsequent stages. Indeed, it is often necessary to consider the power amplifier proper as a complete circuit from the testing point of view. This is because a faulty transistor or component in one stage could reflect fault conditions into the other stages as a result of the d.c. couplings. This possibility is often overlooked when testing in circuits of this kind.

When we reach the point where the signal disappears, therefore, we may have to continue the exercise in terms of voltage measurements at the electrodes of the various transistors, comparing them with the correct values given in the service manual or on the circuit. Complete failure commonly implies significant change in the d.c. conditions, so the fault area should soon be revealed by voltage testing. If a coupling capacitor has failed, then signal tracing should reveal this without much trouble.

More difficult faults to locate are those resulting in low gain, intermittent crackling and distortion at all powers or only at high power. Starting first with distortion, a preliminary test should be that of quiescent current (assuming the amplifier to be 'Class B.' which it usually is in these days of transistors).

If possible, one should avoid connecting a milliammeter in series with the output pair for this test, owing to the resistance of the meter movement. In a circuit like that in Fig 4, the voltage across the emitter resistor (R2) of Tr7 should be measured on a d.c. millivoltmeter. The voltage will be very low because R2 is only 1 ohm! Anyway, suppose we measure 25mV, then by Ohm's law be can calculate that the current is 25mA (1 ohm resistor, since I=V/R, where I is in mA when V is in mV). A current of 25mA is, in fact, a common quiescent current value, and it can be adjusted accurately by a preset such as VR2.

If there is no preset and the current is way out, the values of the resistors associated with this part of the circuit should be measured (also the thermistor, etc., if used). Excessive quiescent current will cause the output transistors to run hot and limit the output

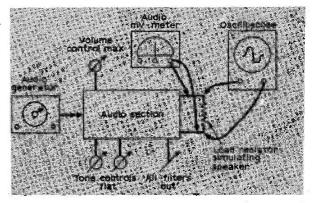


Fig. 5: Equipment set-up required to check the balance of output transistors.

power, while too low a value will precipitate high distortion (crossover distortion) at low volume.

Balancing requires an oscilloscope to monitor the output sinewave fed in from an audio generator. The setup is shown in Fig. 5 where is will be seen that the speaker is replaced by a load resistor of suitable power rating (to match the steady-state power of the amplifier). The volume control should be fully advanced, and the generator signal increased until the tips of the displayed waveform commence to clip. When the balancing is correct both tips will clip together. If one clips before the other then the balancing preset (such as VR1 in Fig. 4) should be adjusted for symmetrical clipping. The waveforms are shown in Fig. 6.

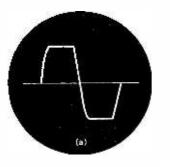
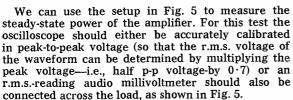


Fig. 6(a) Symmetrical clipping of the waveform when the gain is advanced indicates correct balance.

(b)

Fig. 6(b) Output waveform with an unbalanced stage.



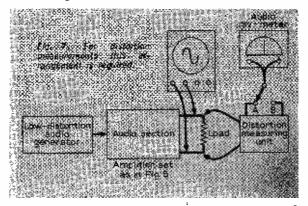
The load voltage just prior to the clipping point should be noted, and the steady-state power in watts is equal to the load voltage squared divided by the load value in ohms (i.e., $W = V^2/R$). The power is usually taken at 1kHz, but the power bandwidth can be appraised by noting the reduction in power at the low and high frequency ends of the spectrum.

Low power, but good waveshape often signifies low supply rail voltage, and this can result from low mains voltage or incorrect mains voltage tapping.

DISTORTION TEST

Finally, total harmonic distortion testing. For this we require either a complete distortion factor meter or the set-up in Fig. 5 with a distortion measuring unit which, as was noted in the Test Instrument supplement, April 1971, is a 'notch filter' which tunes out the fundamental frequency of the output signal, leaving only the harmonic components, the voltage sum of which is compared in ratio with the voltage of the fundamental frequency across the load, to give the dB or percentage total harmonic distortion.

The complete set-up is shown in Fig. 7. The voltage across the load is first measured at the required power (use the oscilloscope for maximum power test to indicate the clipping level) with the switch in postion A, then switching to position B, allowing the audio millivoltmeter to read the voltage of the summed harmonics. For this to happen, however, the distortion measuring unit needs to be very accurately tuned and balanced for the **least** indication on the audio millivoltmeter, which requires the range of the millivoltmeter progressively to be stepped down. The total harmonic distortion is the voltage ratio between the readings at switch postions A and B.



For example, if voltage A is 9V (which across 8 ohms would correspond to about 10W) and voltage B is 90mV, the ratio would be 100:1 or 40dB, corresponding to 1% t.h.d. With 9mV at B it would represent a ratio of 1000:1 or 60dB, corresponding to 0.1% t.h.d.

Remember, though, that a simple distortion measuring set 'notches out' only the fundamental, leaving wideband noise as well as t.h.d. components, so the distortion readout will include noise. Residual hum is generally removed by a high-pass filter in the test set-up, but such filtering, of course, cannot be used for distortion measurements at low-frequency for obvious reasons.

END OF PART SIX

Back Numbers Important Announcement

We regret to inform readers that owing to the closure by the Company of the department concerned it will no longer be possible to supply back numbers of **Practical Wireless** and **Television**.

To ensure obtaining regular copies of these magazines readers are strongly urged to place a regular order with their local newsagent, or to take out an annual postal subscription.

Reference to past issues of the magazines may sometimes be obtained at certain public libraries who may hold bound volumes. A few libraries are said to offer a photostat service. Alternatively, we are always willing to insert a free request for specific back numbers in our "CQ" column which appears in most issues.



N⁰, not that Ted. Henry is not taking the lid off political chicanery—leave that to out-of-office politicians. TED, or, more properly, T.E.D., is a beautiful phrase I heard during a broadcast interview recently. That ubiquitous character, A. Spokesman, was defending the Post Office in the face of polite but relentless attacks from the BBC, who were following up listeners' complaints about crossed lines.

'How does it happen,' came the suave question, 'that we pick up our telephone and find our conversation interrupted by a pair of drooling lovers or an irate customer who has lost his laundry?'

Ah, well, our indefatigable spokesman countered, that is probably caused by a Transient Equipment Defect. Isn't that lovely? Imagine, Joe, when wifey grumbles about the birdies that chirp unwantedly during her 'Housewives' Choice', being able to say airily: "Can't do much about that, I'm afraid. That's TED."

The point is that the faults are not just short-lived, ephemeral, unidentified phenomena—giving them a name has fixed them, preserved them for posterity, entered them on the list of recognisable bugbears.

Wireless work is full of such



We cheer when we get a half-scale reading . . .

instances; for example, the intermittently blowing fuse. We pursue diligent tests before replacing the offending glass tube. We measure resistance, cheer when we get a half-scale reading, then remember that we intended to change the battery of the meter last time we used it.

Baffled by the lack of shortcircuits we decide to beat the brute by checking the current. Meter in series—ah, which way round? Switch on. No, the other way around: *is* that pointer bent? But the reading is no greater than expected, and certainly less than the rating of the fuse, which we refit. Perhaps it is the result of overheating. We cover the now smug-looking set with our jacket and wait.

It is a bad habit of some manufacturers to let dirty great heatsinks jut out at the back, and the fallacy about cool-running transistors is shattered as we smell singeing Harris Tweed and realise we have a Class A output to contend with.

Touching the fins to confirm our nose, we snatch fingers back with a howl and, sucking them, muse that the problem could not have been temperature at all. And the fuse still holds.

Obviously, TED has been at work. He is only waiting for us to settle comfortably with a book and a glass at our elbow before blowing the fuse again. Or, more likely, lulling us into a false sense of security, will allow us the whole evening's listening, then, while we sleep through the small CQ hours, will sneakily contract that vulnerable thread of wire again so that rising betimes, we gaily throw the switch and get: silence!

TED is no fool. That Post Office bloke urged 'sweet reasonableness' upon subscribers when their lines were crossed. The inference was that if we pretended not to notice the fault it would go away. Any PW reader could have told him that is the wrong approach.



No. 84

Ted

and Lid

..., and we snatch our fingers back with a howl.

How many times have you had a piece of equipment propped up on the bench, festooned with test leads, bedizened with warning lights, burbling away in the corner of your mind, and leaped into the breach as the expected fault occurs, only to be greeted with a return to normality the moment you touch the chassis?

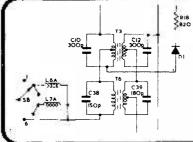
Tracking down TED is the work of an expert. One tests piecemeal. One uses not sweet reasonableness, but bitter reason. Diagnosis, as our soon-to-finish series on Servicing has emphasised, is the secret of success. And for diagnosis, where TED is concerned, read inspired guesswork. This is where LID comes in. LID is one of those contractions I am no longer likely to forget, since talking to Bob Stockwell, publicity manager of Bang & Olufsen, the day after an exhausting exhibition.

'How do you manage to organise it?'

'It's not easy,' he said, 'but we begin by catering for the Largest Imaginable Disaster and take it from there....'

That is the secret of dealing with TED. Put the LID on him: when the fuse blows, imagine the worst that could happen and assume that it has! But don't try to telephone the repairman—you will probably get the laundry company again.





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Step by step we take you through the apparent intricacies of electronics and show you how easily you can master the subject. Write for the brochure which shows, in full detail, how you can do it.



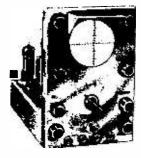
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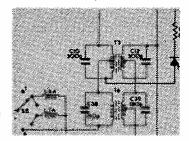


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- Simple counter
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This course has been designed to help even those who are complete newcomers to electronics. No Maths needed !



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THE BROADCAST BANDS Malcolm Connah

MONTHLY **NEWS FOR** DX LISTENERS

THE first report this month comes from a new reporter D. A. Hairon from Jersey in the Channel Islands. The equipment used was a Codar CR70A receiver, a 50ft long wire and a 19 metre band dipole. This combination enabled him to hear the following stations:-

- 9480 Radio Peking in English at 0130.
- 9625 CBC, Canada in English at 0200.
- 11710 R. Nacional de Espana in Spanish at 0130.
- 11720 CBC, Canada in English at 0200.
- 11730 R. Nederland, Bonaire, English at 0200.
- 11745 HCJB, Quito, Ecuador, English at 0000.
- 11775 TWR, Bonaire in Spanish at 0215.
- 11810 RAI, Rome in English at 0100.
- 15195 R. Ankara, Turkey, Turkish at 1600.
- 15250 RSA, South Africa in French at 1856.
- 15300 HCJB, Quito, Ecuador in Spanish at 2145, and in Portuguese at 0000.
- 15305 SBC, Switzerland in English at 1315.
- 15345 Radio Kuwait in English at 1630.
- 15355 ABC, Australia in English at 0530
- 17655 R. Cairo, U.A.R. in English at 1730.
- 17715 ABC, Australia in French at 2315.
- 17810 R. Nederland, Bonaire in Spanish at 2130.
- 17830 R. Lebanon in English at 1830.
- 17825 NHK, Japan in English at 0800.
- 21525 WNYW, U.S.A. in English at 1800.
- 21640 R. Nederland, Bonaire, English at 1830.
- 25730 R. Norway in English at 1200.

Ray Warren of Bury St. Edmunds has used his Koyo 1661 receiver to good effect and heard the following:-

- 3980 Voice of America in English at 1720.
- 5970 R. Free Europe, Germany at 1450.
- 6075 WNBC, New York, news at 1630.
- 6085 R. Nederland in English at 0930.
- 9030 R. Peking in English at 2050.
- 9390 R. Tirana. Albania in English at 1530.
- 9750 SBC, Switzerland, news in English, 1540.
- 15300 HCJB, Quito, Ecuador, English news, 1410.

Peter Hall of Pocklington, Yorkshire, has a Meridian 10 transistor superhet and a 50 foot aerial which enabled him to hear:-

- 6070 R. Sofia, Bulgaria in English at 1930.
- 6130 HCJB, Quito, Ecuador in English at 0800.
- 7275 RAI, Rome in English at 1940.
- 9550 Radio Finland in English at 1815.
- 9580 BBC, Ascension Island relay at 2115.
- 9620 R. Belgrade, Yugoslavia at 1800.
- 9625 Israel B.C. in English at 2045.
- 9630 R. Trans Europe, Portugal at 2100.
- 9700 R. Sofia, Bulgaria in English at 2140.
- 11620 AIR, Delhi in English at 2010.
- 11672 R. Pakistan in English at 2030.

- 11755 R. Finland in English at 0930.
- 11915 BBC, Cyprus relay at 1645.
- 11965 TWR, Monaco in English at 1700.

Ross Pullen of Crawley used his Murphy A72 communications receiver and a 100 foot long-wire to hear some interesting stations:-

- 6015 VOA, Rhodes, music and talk at 2200.
- 6145 R. Nacional, Brazil in Spanish at 2245.
- 9510 R. Noumea, New Caledonia, French at 0910.
- 9605 HCJB, Quito, Ecuador, English at 0730.
- 9745 R. Baghdad, Iraq in English at 2015.
- 9775 R. Diffusion-Television Nationale Congolaise, African music at 2200.
- 11672 R. Pakistan, English at 2002.
- 11965 Rwanda relay of Deutsche Welle at 2037.
- 17705 R. Havana, Cuba, Spanish news at 2015.

F. Wall of Walthamstow, London, E.17, has a homebrew 5-valve superhet and a 20 foot wire aerial, this equipment enabling him to hear:-

- 6035 R. Warsaw with Polish news at 1620.
- 6070 R. Sofia, Bulgaria in English at 1930.
- 9912 AIR, Delhi English news at 2000.
- 9630 R. Sweden 'Calling DXers' at 1230.
- 9660 R. Trans Europe in English at 1250.
- 9833 R. Budapest, Hungary, requests at 1945. 11735 R. Belgrade, Yugoslavia, English at 1530.
- 11740 ABC, Australia at 1530.
- 11755 R. Finland in English at 1500.
- 11765 ABC, Australia, sports programme at 0830.
- 11955 BBC, Malaysian relay at 1815.
- 15325 CBC, Canada, sports news at 1200.
- 21460 HCJB, Quito, Ecuador sign-on in French at 2015.
- 21535 RSA, South Africa, English news at 0930. 21595 CBC, Canada with news at 1910.
- 25790 RSA, South Africa, English news at 1500.

Stephen Mathews of Hull with his Bush 4-valve domestic receiver and 90 foot loft aerial heard the following:-

9660 YVLM Radio Rumbos, Venezuela at 0320.

- 9833 R. Espana Independiente, clandestine at, 2230.
- 11765 ZYB8, R. Diffusion, Brazil at 0220.
- 11795 WINB, Red Lion, U.S.A. English at 2100.
- 11815 ZYW24, R. Brazil Central at 0045.
- 15260 FEBA, Seychelles testing at 1700.
- 17890 BED-40, Taiwan, weak at 0330.
- 21690 WIBS, Grenada in English at 1945.
- 21720 R. Ghana in English at 1445.

Reports should arrive by the 15th of the month and be addressed to the author at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.



H IGH noise levels have been the main complaint this past month with many people confessing to giving up and going off to watch television (sacrilege!). Despite these conditions there was still a large number of listeners who persisted, despite the noise, and managed to log some worthwhile DX as reward for their labours. Just goes to show, while you are busy moaning about the QRM and QRN, someone else is just getting on with it and working half the world.

News that G5ZT has received slow-scan television pictures from FG7XT in Guadeloupe makes you think. Many people would be only too pleased to hear an FG7 on c.w. **John Rimmer** (Deal) warns that VS9MF is on from Gan—a small island in the Maldive group of islands. Listen on 14 and 21MHz; sorry, no specific freqs given.

No activity reported on ten metres this month and it rather looks as though everyone has migrated to 14MHz. This latter band is an early morning/late night affair nowadays with everything dying out from mid-morning to late afternoon. On fifteen metres the same remarks apply, although occasionally South American stations appear during the daytime. For those who would like to persist on 28MHz, suggested time for a listen is from 1400 to 1600hrs.

Incidentally, this is one of the best bands on which to play with aerials since these normally work out to a very reasonable size. A quarter wave vertical is only around the 8ft. 6in. mark, exact size depending upon frequency since ten metres is one of the widest bands we have among those l.f. of 144MHz. If sunspot counts are any indication, then September would be a very good time for ten metres

A few contests this month just to get your ears in trim for the long winter evenings. September 11 and 12, WAE phone contest; 12, 3 5MHz Field Day; October 2 and 3, u.h.f./s.h.f. Field Day; 2 and 3, Oceania contest-a chance to bag a nice list of VK/ZL callsigns; 2 and 3, IARU Region 1 u.h.f./s.h.f. contest. Don't forget the Cumulative Activity on 70MHz from October 9 to December 30. This is our "use or lose" band and all support is very welcome, if not essential. You might try listening for some of the RAEN activity on four metres. The Radio Amateur Emergency Network has numerous "local" groups and the favourite band is 70MHz. How about listening out for your local group, or pérhaps even joining them? Name and address of the Secretary of RAEN is; E. Bassett, 57 Upper St Helen's Road. Hedge End, Southampton, SO3 4LG.

Logs

A solitary log for two metres arrived this month from **David Sheekay** (Ashford). Gear is a superregen using three transistors and an i.c. audio amplifier, homebrew, and antenna is described as "telescopic". Stations logged include; G2JF, G2VB, G3EFX/P, G3JXN, G3LTF, G3TDP/M, G3UNT,

THE AMATEUR BANDS David Gibson, G3JDG

Frequencies in kHz • Times in GMT

G4AAR, G4AAZ/M, G5OX, G8BYC, G8CCO, G8EBV, G8ELW and F0UJ. What about it all you converter and superhet types—you can't let a super-regen show you up, or can you?

Richard Mortimore (Cardiff), is now up to 20 w.p.m. at c.w. and says that he recommends all s.w.l.s to learn morse. Small sample from his list of c.w. pickings on 21MHz includes; CR6ID, JA3IL, KZ5BB, PY2EYE, PY4UK, W4NZK, WB4QVQ.

Mark Marsden (Ilford) uses a cassette tape recorder which has its input wired in parallel with the headphones. He claims this is very useful for checking calls heard against calls logged afterwards. Seems a very good idea. Calls heard, checked, recorded and logged on 14MHz include signals from; CE5ID, JA3JH, KV4HE, KZ5JF, PY2ZAC, PY4AEB, SV0WZ, VK3ART, VK4PJ, VP2AA, WA1DYR/MM, WB6YIY/MM, YV5CPA, YN2SC, ZL5ISJ, 4X4JU, 4X4OC, 9Q5IA.

"My uncle has given me all the PWs he can spare", writes **Kevin Davies.** Receiver is an R107 and the antenna 120ft. l.w. The following were in evidence on 14MHz s.s.b.; CT1ZE, EA3KO, EA4LM, EL2C, EL4A, HR3VSJ, JA1VK, K1WPS/CT1, LU4DMG, W1CMG/P1, W1MYA/MM, WA5KHM, WB6YIY, 4U1ITU, 4X4HT.

Paul Newman is at a new QTH which is some 680ft. a.s.l. Antenna at present is an 80ft. end fed and the receiver a modified BC348R ex G3HXS. With this gear, and from his lofty perch at Aylesbury, Paul bagged these on 14MHz; OD5BV, PY6AT, PY8LM, (Sao Luis Island), WA5STY/P/YV5 (Caracas) 4X4BL, 4X4GV, 5Z4KL, 9K2AL.

Jim Martin says that as it's the start of his holidays he thought he'd send in a report. (May the sun shine on your antenna, Sir.) His recentlypurchased HA500 is nourished with an inverted V and rewards on 14MHz were; CR6EF, CR7IK, EL2CY, FP8CS, HC6CB, JA3IXL, JA4EDY, JA6FUV, JY9XL, KG4AN, KH6FP, KH6BB (Kure Island), KL7DIV, KR6US, KV4AD, KZ5AO, LG5LG, MP4MBC, PY1FN, PJ2MI, PZ1AD, PZ1AW, TG9GF, VK3JS, VP2MO, VQ9R, VS5CB, VS9MT, VU2BEZ, YV5EC, ZE1DO, ZP5GU.

On 3.5MHz c.w. it was UA1YUZ. Up on 7MHz, signals came from CX6CG, LX1BJ, PYCVX, PZ1AX, UW3NG, ZB2A, and ZP5AR. Gear is a JR500SE with 132ft. end fed. Operator's ears belong to Julian Iredale (Llandudno), and up on 14MHz s.s.b. Julian heard CR6GA, HV3SJ, OG5A, (Finnish Field Day prefix), PY2ZAC, PY4ATL, PZ1DL, TI2J, 5A4DL.

P. McKay advises that he never notices any logs for 7MHz in the Amateur Bands, and sends in a list of G-stations heard on this band. Well, the G's must have been working someone, so how about a listen this month?

Logs, in alphabetical order please, to arrive by the 15th of the month to: 12, Cross Way, Harpenden, Herts.





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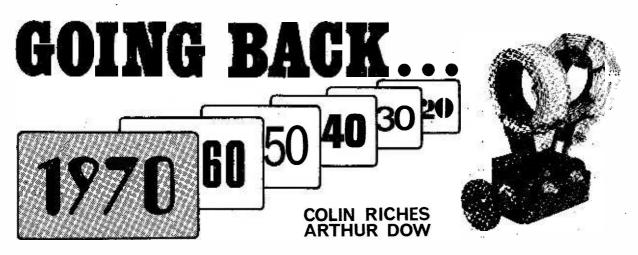
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D. C. TIBBLES & CO. PRI. 0-1 MINIATURE TRANSFORMER $2^{"} \times 1\frac{1}{2}^{"} \times 1\frac{1}{4}^{"}$	00.110-990-9		8 5 0.1977	90.4		\$4.50	each
D. C. TIBBLES & CO. PRI. 0-1 MINIATURE TRANSFORMER	00-110-220-5		8 5 0.1977	. 20A mA 30V 10	 0mA	£6-50	
D. C. TIBBLES & CO. PRI. 0-1 MINIATURE TRANSFORMER	8. PRI. 240 PRI. 240 PRI. 240 PRI. 240	V SEC 0 V SEC 0 V SEC 0	C. 12V 2V 100 , 20V, ,18,36,	. 20A mA 30V 10	 0mA	£6-50	
D. C. TIBBLES & CO. PRI. 0-1 MINIATURE TRANSFORMER 2"×11*×11* SHROUDED TRANSFORMER	00-110-220-9 S. PRI. 240 PRI. 240 PRI. 240 S. PRI. 0-1 PRI. 0-2	ORMER 240V. SE V SEC 1 V SEC 0 V SEC 0, 20-240V 40V SEC	C. 12V 2V 100 , 20V, .18, 36, SEC. 2 26V 1	. 20A mA 30V 10 , 0, 18, 3 4V 1A A		£6.50 nA es 65p £1.25 £1.60	each each each
D. C. TIBBLES & CO. PRI. 0-1 MINIATURE TRANSFORMER 2"×1½"×1¼"	8. PRI. 240 PRI. 240 PRI. 240 PRI. 240	DRMER 240V. SE V SEC 1 V SEC 0 V SEC 0, 20-240V 40V SEC 0 V SEC 0	C. 12V 2V 100 , 20V, , 18, 36, SEC. 2 26V 1 , 74, 14	. 20A mA 30V 10 , 0, 18, 3 4V 1A	 0mA 36 100n All Typ 	£6-50	each each each each
D. C. TIBBLES & CO. PRI. 0-1 MINIATURE TRANSFORMER 2"×1\\$"×1\\$" SHROUDED TRANSFORMER C-CORE TRANSFORMERS.	00-110-220-5 S. PRI. 240 PRI. 240 PRI. 240 S. PRI. 0-1 PRI. 0-2 PRI. 240 PR. 0-11	DRMER 240V. SE 1V SEC 1 V SEC 0 V SEC 0, 20-240V 40V SEC 0 V SEC 0 0-200-220	$\begin{array}{c} \textbf{S} \\ \textbf{C}, 12V \\ 2V 100 \\ , 20V \\ , 18, 36 \\ \textbf{SEC}, 2 \\ 26V 1 \\ , 7\frac{1}{2}, 10 \\ 0.240 \\ \textbf{S} \end{array}$. 20A mA 30V 10 , 0, 18, 3 4V 1A	 0mA 36 100n All Typ 	£6.50 nA es 65p £1.25 £1.60 £1.20	each each each each
D. C. TIBBLES & CO. PRI. 0-1 MINIATURE TRANSFORMER 2"×1\\$"×1\\$" SHROUDED TRANSFORMER C-CORE TRANSFORMERS.	00-110-220-5 S. PRI. 240 PRI. 240 PRI. 240 S. PRI. 0-1 PRI. 0-2 PRI. 240 PR. 0-110 ALARM	DRMEF 240V. SE 240V. SEC 1 V SEC 0 V SEC 0, 20-240V 40V SEC 0 0-200-220 BELLS	C. 12V 2V 100 , 20V, .18, 36, SEC. 2 26V 1 , 7½, 10 0.240 S	. 20A mA 30V 10 , 0, 18, 3 4V $\frac{1}{2}$ A A 5V 1A EC 6.3	 0mA 36 100n All Typ 	£6.50 nA es 65p £1.25 £1.60 £1.20	each each each each each
D. C. TIBBLES & CO. PRI. 0.1 MINIATURE TRANSFORMER 2"×1½"×1½" SHROUDED TRANSFORMERS. HEATER TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A	00-110-220-5 S. PRI. 240 PRI. 240 PRI. 240 S. PRI. 0-1 PRI. 0-2 PRI. 240 PRI. 0-2 PRI. 240 COMPANY CO	DRMER 240V. SE 1V SEC 1 V SEC 0 20-240V 40V SEC 0 00-200-220 BELLS 1145 200/	C. 12V 2V 100 , 20V, ,18,36, SEC. 2 26V 1 ,7 ¹ / ₂ , 10 0-240 S 250V a	. 20A mA 30V 10 , 0, 18, 3 4V $\frac{1}{2}$ A A 5V 1A EC 6.3	 0mA 36 100n All Typ 	£6.50 nA es 65p £1.25 £1.60 £1.20 £1.00 £3.50	each each each each each each
D. C. TIBBLES & CO. PRI. 0.1 MINIATURE TRANSFORMER 2"×1½"×1½" SHROUDED TRANSFORMERS. HEATER TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A	00-110-220-5 S. PRI. 240 PRI. 240 PRI. 240 S. PRI. 0-1 PRI. 0-2 PRI. 240 PRI. 0-2 PRI. 240 COMPANY CO	DRMER 240V. SE 1V SEC 1 V SEC 0 20-240V 40V SEC 0 00-200-220 BELLS 1145 200/	C. 12V 2V 100 , 20V, ,18,36, SEC. 2 26V 1 ,7 ¹ / ₂ , 10 0-240 S 250V a	. 20A mA 30V 10 , 0, 18, 3 4V 1A 5V 1A EC 6-3 c.	 0mA 36 100n All Typ V 3A 	£6.50 nA es 65p £1.25 £1.60 £1.20 £1.00 £3.50	each each each each each each
D. C. TIBBLES & CO. PRI.0-J MINIATURE TRANSFORMER 2"×1½"×1½" SHROUDED TRANSFORMER C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,458 SIFAM MOVING COIL 100,458 SIFAM MOVING COIL 100,458	00-110-220-5 \$. PRI. 240 PRI. 240 PRI. 240 \$. PRI. 0-1 PRI. 0-2 PRI. 240 PR. 0-11 ALARM GRO Type 4 PANEL N 26 4 ³ / ₄ ×4" Size 3 ³ / ₄ ×3" OµA Edgewi	DRMEF 240V. SE V SEC 1 V SEC 0, 20-240V SEC 0, 20-240V SEC 0 0-200-220 BELLS BELLS 1145 200/ 1ETER ise 2¼* sc	RS C. 12V 2V 100 , 20V, ,18, 36, SEC. 2 26V 1 , 7 ¹ / ₂ , 16 0-240 S 250V a S ale	. 20A mA 30V 10 , 0, 18, 3 4V $\frac{1}{2}$ A 3V 1A EC 6.3 c.	 36 100m All Typ V 3A 	£6.50 nA es 65p £1.25 £1.60 £1.20 £1.00	each each each each each each each
D. C. TIBBLES & CO. PRI. 0.1 MINIATURE TRANSFORMER 2"×1½"×1½" SHROUDED TRANSFORMER C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COLL 1mA Si SIFAM MOVING COLL 1mA Si SIFAM MOVING COLL 100,4AS TAYLOB MOVING COLL 50-0-5 ERNEST TURKER 270° MOVEN	00-110-220-5 \$. PRI. 240 PRI. 240 PRI. 240 \$. PRI. 0-1 PRI. 0-2 PRI. 240 PR. 0-11 ALARM GRO Type 4 PANEL N 26 4 ³ / ₄ ×4" Size 3 ³ / ₄ ×3" OµA Edgewi	DRMEF 240V. SE V SEC 1 V SEC 0, 20-240V SEC 0, 20-240V SEC 0 0-200-220 BELLS BELLS 1145 200/ 1ETER ise 2¼* sc	RS C. 12V 2V 100 , 20V, ,18, 36, SEC. 2 26V 1 , 7 ¹ / ₂ , 16 0-240 S 250V a S ale	. 20A mA 30V 10 , 0, 18, 3 4V $\frac{1}{2}$ A 3V 1A EC 6.3 c.	 36 100m All Typ V 3A 	£6.50 nA £1.25 £1.60 £1.20 £1.00 £3.50 £3.25 £3.00 £4.25	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J MINIATURE TRANSFORMER 2"×1½"×1¼" SHROUDED TRANSFORMER. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,40.8 SIFAM SIFAM SIF	.00-110-220-5 S. PRI.240 PRI.240 PRI.240 PRI.240 S. PRI.02 PRI.02 PRI.02 PRI.04 PRI.04 <	DRMEF 240V. SE VV SEC 1 VV SEC 0, 20-240V 40V SEC 0, 20-240V 40V SEC 0, 0-200-22C BELLS 1145 200/ 1ETER	X C. 12V 2V 100 , 20V, , 18, 36, SEC. 2 26V 1 , 7½, 10 -240 S 250V a S ale alar 0-5	. 20A mA 30V 10 , 0, 18, 3 4V $\frac{1}{2}$ A A SV 1A EC 6·3 	 0mA 36 100m All Typ V 3A nA	£6.50 nA wes 65p £1.25 £1.60 £1.20 £3.50 £3.50 £3.00 £4.25 £3.00	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J MINIATURE TRANSFORMER 2"×1½"×1¼" SHROUDED TRANSFORMER. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,40.8 SIFAM MOVING COIL 100,40.8 SIFAM MOVING COIL 100,40.8 SIFAM MOVING COIL 100,40.8 ENEST TURNER 270° Moven 24" Dia.	00-110-220-5 S. PRI. 240 PR. 241	DRMEF 240V. SE V SEC 1 V SEC 0, 20-240V 40V SEC 0, 20-240V 40V SEC 0 0-200-220 BELLS 145 200/ 1ETER (se 2¼* sc erro Circu 3 zero C	C . 12V 2V 100 , 20V, , 18, 36, SEC. 2 26V 1 , 7 ¹ , 1 ¹ 0-240 S 250V a S ale llar 0-5 	. 20A mA 30V 10 0, 18, 3 4V 1A EC 6.3 1-0-1r	 0mA 36 100m All Typ V 3A nA nA 	£6.50 nA ess 65p £1.25 £1.60 £1.20 £1.20 £1.20 £1.20 £1.20 £1.20 £1.20 £1.20 £1.25 £1.60 £1.25 £3.00 £1.25 £3.00 £3.25 £3.00 £4.25 £3.00 £4.25 £3.00 £4.25 £3.00 £4.25 £3.00 £4.25 £3.00 £4.25 £3.00 £4.25 £3.00 £4.25 £3.00 £4.25	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J MINIATURE TRANSFORMER 2"×1½"×1½" SHROUDED TRANSFORMERS C-CORE TRANSFORMERS, HEATER TRANSFORMERS, Suitable for Burglar Alarms, A SIFAM MOVING COIL 1mA SI SIFAM MOVING COIL 1mA SI SIFAM MOVING COIL 1mA SI SIFAM MOVING COIL 1mA SI SIFAM MOVING COIL 1mA SI ENTER TURKER 270° MOV 34" Dia. ERNEST TURKER 270° MOV 34" Dia.	00-110-220-5 S. PRI. 240 PRI. 240 PRI. 240 S. PRI. 0-2 PRI. 240	DRMEF 240V. SE: VV SEC1 VV SEC0 20-240V 440V SEC0 0-200-220 BELLS 1145 200/ 1ETER	C . 12V 2V 100 , 20V, , 18, 36, SEC. 2 26V 1 , 7 ¹ , 1 ¹ 0-240 S 250V a S ale llar 0-5 	. 20A mA 30V 10 0, 18, 3 4V 1A EC 6.3 1-0-1r	 0mA 36 100m All Typ V 3A nA nA 	£6.50 nA es 65p £1.20 £1.20 £3.50 £3.50 £3.25 £3.00 £4.25 £3.90	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-) MINIATURE TRANSFORMER 2"×1½"×1½" SHROUDED TRANSFORMERS HEATER TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Aiarms. A SIFAM MOVING COLL 100.AS SIFAM MOVING COLL 100.AS TAYLOR MOVING COLL 50-0-5 ERNEST TURKER 270° MOV 34" Dia. ERNEST TURKER 270° MOV 34" Dia. SANGAMO-WESTON. 150.4AF SANGAMO-WESTON. 150.4AF We also have a large number o	00-110-200-5 S. PRI. 240 PRI. 240 PRI. 240 PRI. 240 PRI. 0-2 PRI. 0-2 PRI. 240 PR. 0-11 ALARM GRO Type 4 PANEL N 22 4 ³ / ₂ ×4" NO ₄ A Edgewi tent/Centre 2 memor/Centre 2 MATED METR	DRMEF 240V. SE: VV SEC 1 VV SEC 0, 20-240V 40V SEC 0, 20-240V 40V SEC 0, 0-200-220 BELLS 1145 200/ 11ETER	C . 12V 2V 100, 20V, 118, 36, SEC. 2 26V 1, 7 ¹ / ₂ , 1; -240 S 250V a S ale lar 0-5 F.S.D 	$\begin{array}{c} 20A \\ mA \\ 30V 10 \\ 0, 18 \\ 14V \frac{1}{2}A \\ A \\ 5V 1A \\ EC 6 \cdot 3 \\c. \\ \\ 1 - 0 - 0 \cdot 5 \\ 1 - 0 - 1r \\ \\ 4\frac{1}{2}^{v} \times 3 \\ \end{array}$	 36 100m 36 100m All Typ V 3A A A A A A 	£6.50 1A 1A 125 £1.25 £1.20 £1.20 £3.50 £3.50 £3.25 £3.00 £4.25 £3.00 £4.25 £3.90 £4.25	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J MINIATURE TRANSFORMER 2"×1½"×1¼" SHROUDED TRANSFORMER C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,4A S SIFAM MOVING COIL 100,4A S SANGAMO-WESTON ILLUKIN SANGAMO-WESTON SO,4A F.3	00-110-200-5 S. PRI. 240 PRI. 240 PRI. 240 PRI. 240 PRI. 0-2 PRI. 0-2 PRI. 240 PR. 0-11 ALARM GRO Type 4 PANEL N 22 4 ³ / ₂ ×4" NO ₄ A Edgewi tent/Centre 2 memor/Centre 2 MATED METR	DRMEF 240V. SE: VV SEC 1 VV SEC 0, 20-240V 40V SEC 0, 20-240V 40V SEC 0, 0-200-220 BELLS 1145 200/ 11ETER	C . 12V 2V 100, 20V, 118, 36, SEC. 2 26V 1, 7 ¹ / ₂ , 1; -240 S 250V a S ale lar 0-5 F.S.D 	$\begin{array}{c} 20A \\ mA \\ 30V 10 \\ 0, 18 \\ 14V \frac{1}{2}A \\ A \\ 5V 1A \\ EC 6 \cdot 3 \\c. \\ \\ 1 - 0 - 0 \cdot 5 \\ 1 - 0 - 1r \\ \\ 4\frac{1}{2}^{v} \times 3 \\ \end{array}$	 36 100m 36 100m All Typ V 3A A A A A A 	£6.50 1A 1A 125 £1.25 £1.20 £1.20 £3.50 £3.50 £3.25 £3.00 £4.25 £3.00 £4.25 £3.90 £4.25	each each each each each each each each
D. C. TIBBLES & CO. PRI.0.) MINIATURE TRANSFORMER 2"×1½"×1¼" SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,40.8 SIFAM MOVING SIFAM	100-110-220- S. FRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI	DRMEF 240V. SEC V SEC 1 V SEC 1 V SEC 0 20-240V V SEC 0 0-200-220 BELLS 1445 200/ 4ETER 24" sc erro Circu 3 zero C CR 500 µLA "eters of d	(S C. 12V 2V 100 2V 100 , 20V, 18, 36, SEC. 2 26V 1 , 7½, 14 0-240 S 250V a S ale llar 0-5 ircular F.S.D 	. 20A mA 30V 10 , 0, 18, ; 4V 1A EC 6.3 c. -0-0.5 r 1-0-1r t shape	 36 100m All Tyr V 3A nA aA s/sizes/	£6.50 hA es 65p £1.25 £1.60 £1.20 £1.00 £3.50 £3.25 £3.25 £3.25 £4.25 £3.90 £4.25 £3.90 £4.25 £3.90	each each each each each each each each
D. C. TIBBLES & CO. PRI.0.) MINIATURE TRANSFORMER 2"×1½"×1¼" SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,40.8 SIFAM MOVING SIFAM	100-110-220- S. FRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI 240 PRI 250 PRI	DRMEF 240V. SEC V SEC 1 V SEC 1 V SEC 0 20-240V V SEC 0 0-200-220 BELLS 145 200/ 162 145 200/ 162 145 200/ 162 145 200/ 162 163 165 24" sc. 165 165 165 165 165 165 165 165 165 165	(S C. 12V 2V 100 2V 100 , 20V, 18, 36, SEC. 2 26V 1 , 7½, 14 0-240 S 250V a S ale llar 0-5 ircular F.S.D 	. 20A mA 30V 10 , 0, 18, ; 4V 1A EC 6.3 c. -0-0.5 r 1-0-1r t shape	 0mA 100m 11 Typ V 3A 	\$6.50 nA \$1:25 \$1:60 \$1:20 \$1:	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J. MINIATURE TRANSFORMER 2'×1½'×1¼' SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,45 S SIFAM MOVING SIFAM S SIFAM S SI	100-110-220- S. FRI 240 PRI	DRMEF 240V. SE 240V. SEC 0, V SEC 0, V SEC 0, 20-240V 20-20-20 V SEC 0, 20-20-20 BELLS 1445 200/ 145 200/ 1222 BELLS	S 2V 100 2V 101 2V 101 2V 101 2V 101 SEC 2 2267 1 10-240 S 2250V a 2250V a ale ale ale ale 10-15 F.S.DU Jifferen OR DIS	. 20A mA S0V 10 0, 18, : 4V $\frac{1}{2}A$.c. t shape 	 0mA 100m 11 Typ V 3A 	£6.50 nA ves 65p £1.25 £1.60 £1.20 £1.20 £3.50 £3.50 £4.25 £3.00 £4.25 £3.90 £4.25 £3.90 £4.25 £3.90 £4.75 moven ETC.	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J. MINIATURE TRANSFORMER 2'×1½'×1¼' SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,458 SIFAM MOVING COIL 100,458 SINGLE CHANNEL 700 SINGLE CHANNEL 700 SINGLE CHANNEL MODI SINGLE CHANNEL MO	100-110-220-1 S. FRI 240 PRI 2	DRMEF 240V. SEC V SEC 1 V SEC 1 V SEC 0 20-240V V SEC 0 0-200-220 BELLS 145 200/ 162 145 200/ 162 145 200/ 162 145 200/ 162 163 165 24" sc. 165 165 165 165 165 165 165 165 165 165	S 2V 100 2V 101 2V 101 2V 101 2V 101 SEC 2 2267 1 10-240 S 2250V a 2250V a ale ale ale ale 10-15 F.S.DU Jifferen OR DIS	. 20A mA S0V 10 0, 18, : 4V $\frac{1}{2}A$.c. t shape 	 DmA 0mA 11 Tyr V 3A v 3A s/sizes/ £8:0 £35.0	\$6.50 nA \$1:25 \$1:25 \$1:25 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:20	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J. MINIATURE TRANSFORMER 2'×1½'×1¼' SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,45 S SIFAM MOVING SIFAM S SIFAM S SI	100-110-220-1 S. FRI 240 PRI 2	DRMEF 240V. SE 240V. SEC 0, V SEC 0, V SEC 0, 20-240V 20-20-20 V SEC 0, 20-20-20 BELLS 1445 200/ 145 200/ 1222 BELLS	S 2V 100 2V 101 2V 101 2V 101 2V 101 SEC 2 2267 1 10-240 S 2250V a 2250V a ale ale ale ale 10-15 F.S.DU Jifferen OR DIS	. 20A mA 30V 10 0, 18, ; 4V $\frac{1}{2}A$ 5V 1A EC 6.3 1-0-1r 4 $\frac{1}{2}^{*}\times$: 	 DmA 0mA 11 Tyr V 3A v 3A s/sizes/ £8:0 £35.0	\$6.50 nA \$1:25 \$1:00 \$1:00 \$3:50 \$3:00 \$3:25 \$3:00 \$4:25 \$3:00 \$4:25 \$3:00 \$4:25 \$3:00 \$4:25 \$3:00 \$4:25 \$3:00 \$4:25 \$3:00 \$1:	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J. MINIATURE TRANSFORMER 2'×1½'×1¼' SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COLI 100,458 SIFAM MOVING COLI 100,458 SINGLE CHANNEL 700 ⁻ TINGLE CHANNEL 700 ⁻	100-110-220- S. FRI 240 FRI	DRMEF 240V. SE VV SEC1 VSEC0 VSEC0 VSEC0 20-240V 40V SEC 02-240V 40V SEC 02-240V 40V SEC 0-200-22 BELLS 145 145 22* sc ero Chrou 3 zero C RE500 µA ** <tr< td=""><td>R C. 12V 100 , 20V , 20V 100</td><td>. 20A mA 30V 10 0, 18, ; 4V $\frac{1}{2}A$ 5V 1A EC 6.3 1-0-1r 4$\frac{1}{2}^{*}\times$: </td><td> DmA 0mA 11 Tyr V 3A v 3A s/sizes/ £8:0 £35.0</td><td>\$6.50 nA \$1:25 \$1:25 \$1:25 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:20</td><td>each each each each each each each each</td></tr<>	R C. 12V 100 , 20V , 20V 100	. 20A mA 30V 10 0, 18, ; 4V $\frac{1}{2}A$ 5V 1A EC 6.3 1-0-1r 4 $\frac{1}{2}^{*}\times$: 	 DmA 0mA 11 Tyr V 3A v 3A s/sizes/ £8:0 £35.0	\$6.50 nA \$1:25 \$1:25 \$1:25 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:20 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:25 \$1:20	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J. MINIATURE TRANSFORMER 2'×1½'×1¼' SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COIL 100,458 SIFAM MOVING COIL 100,458 SIGLE CHANNEL 700 SINGLE CHANNEL MODI SINGLE CHANNEL	100-110-320-1 S. FRI 240 PRI. 240	DRMEF 240 V. SE VV SEC 0 VV SEC 0 VV SEC 0 20-240V 40V SEC 0 0 0-200-22 BELLS 1445 200/ 14ETER	R C. 12V 100 C. 12V 100 20V 100 y 20V 103 20V 100 y 20V 103 20V 100 y 20V 107 11 100 y 20V 100 100 100 y 20V 100 100 <td>. 20A mA 30V 10 0, 18, ; A V 1A EC 6.3 </td> <td> OmA Soldon All Typ </td> <td>\$6.50 hA les 657 \$1:25 \$1:</td> <td>each each each each each each each each</td>	. 20A mA 30V 10 0, 18, ; A V 1A EC 6.3 	 OmA Soldon All Typ 	\$6.50 hA les 657 \$1:25 \$1:	each each each each each each each each
D. C. TIBBLES & CO. PRI.0-J. MINIATURE TRANSFORMER 2'×1½'×1¼' SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COLL 100,4AS SIFAM MOVING COLL 100,4AS SINGLE CHANNEL 700 IN SLOCK. SOUND-TO-LIGHT MODI SINGLE CHANNEL S' D. METAL OXIDE RESISTORS. each. 100 off Sip each.	100-110-220- 5. FRI 240 FRI	DRMEF 240V.SE W VSEC 10 V SEC 0 V VSEC 0, 20-240V 40V SEC 0 V SEC 0 0-200-220 BELLS 1145 200/ ABELE 200/ ABELE 200/ ABELE 500/ CHARTER CONTRACTOR SEC 100/ ABELE FC CONTRACTOR SEC 100/ CONTRACTOR SEC 100/ SEC 1	XS C. 12V 100 2.2V 100 , 20V, , 10V, , 10V, <t< td=""><td>. 20A mA 30V 10 30V 10 50V 1A EC 6-3 </td><td> OmA \$61000 All Typ s/sizes/ \$600 \$100 s/sizes/ \$620ES \$600 \$100 </td><td>\$6.50 hA tes 657 \$1.25 \$1.60 \$1.25 \$1.60 \$2.20 \$3.50 \$3.00 \$4.25 \$4.75 \$3.00 \$3.</td><td>each each each each each each each each</td></t<>	. 20A mA 30V 10 30V 10 50V 1A EC 6-3 	 OmA \$61000 All Typ s/sizes/ \$600 \$100 s/sizes/ \$620ES \$600 \$100 	\$6.50 hA tes 657 \$1.25 \$1.60 \$1.25 \$1.60 \$2.20 \$3.50 \$3.00 \$4.25 \$4.75 \$3.00 \$3.	each each each each each each each each
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D. C. TIBBLES & CO. PRI.0.1 MINIATURE TRANSFORMER 2'×1½'×14' SHROUDED TRANSFORMERS. C-CORE TRANSFORMERS. HEATER TRANSFORMERS. Suitable for Burglar Alarms. A SIFAM MOVING COLL 100,4AS SIFAM MOVING COLL 100,4AS SINGLE CHANNEL 700 MOVER SINGLE CHANNEL 700 SINGLE CHANNEL MOI SINGLE CHANNEL SO SINGLE CHANNEL SO SINGLE CHANNEL SO SINGLE CHANNEL MOI SINGLE CHANNEL SO SINGLE CHANNEL SO SINGLE CHANNEL MOI SINGLE CHANNEL SO SINGLE SINGLE SI SIMCONDUND RESISTORS. IN INSTRCE SINGLE SINGLE SINGLE	100-110-220-1 S. FRI 240 FRI	AVV. SEC 440V. SEC V SEC 0 V SEC 0 V SEC 0 V SEC 0 0.20-240V 440V SEC FRE 500/ 145 200/ 145 200	US 22 1200 C.1 22 V 1000 9.00 V. y 20 V. 118, 36, 36, 300 y 30 V. 118, 36, 300 y 30 V. 118, 300 y 30 V. <td>204 mA S0V 10, 0, 8, 1 S0V 10, 1</td> <td></td> <td>\$6-50 hA \$1:25 \$1:25 \$1:20 \$2:50 \$1:20 \$2:50 \$1:20 \$1:20 \$2:50 \$1:20 \$2:50 \$1:20 \$2:50 \$2:50 \$1:20 \$2:50 \$2:50 \$1:20 \$2:50 \$1:20</td> <td>each each each each each each each each</td>	204 mA S0V 10, 0, 8, 1 S0V 10, 1		\$6-50 hA \$1:25 \$1:25 \$1:20 \$2:50 \$1:20 \$2:50 \$1:20 \$1:20 \$2:50 \$1:20 \$2:50 \$1:20 \$2:50 \$2:50 \$1:20 \$2:50 \$2:50 \$1:20 \$2:50 \$1:20	each each each each each each each each



M ANY readers have written to ask us if we can suggest any gramophone records that may help them to bring alive even more vividly the days of vintage radio. After a certain amount of research we have discovered that Fountain Press, the book publishers, have now announced a new "Fountain" record label of Vintage Jazz music. The first 12in. LP releases feature historic jazz and blues recordings but future plans include Music Hall, Opera and other music of the "Good Old Days".

The records are housed in attractive double sleeves which open, displaying an authentically documented note section with accurate historical background information and photographs.

The first two releases are: "Original Dixieland Jazz Band and the Louisiana Five"—from the extremely rare Aeolian sessions of 1917 (catalogue number FJ-101). "Ladds Black Aces—Vol. 1"—the first 16 sides recorded by another pioneer white jazz band during 1921-1922 (Cat. No. FJ-102). Following soon are "Naylor's Seven Aces" (c 1923-4). "Jelly Roll Morton" 1923-4 piano solos. "Lovie Austin's Blues Serenaders" (1924-6), and "Annette Hanshawe. Vol. 1"—Miss Hanshawe enjoyed public acclaim as a singer of popular songs in the 1920's and 1930's and her records are great collectors' items today.

These records are available from record shops at $\pounds 2 \cdot 10$ or direct at $\pounds 2 \cdot 25$ which includes post and packing, from Fountain Records, 46 Chancery Lane, London WC2A 1JU.

If you have any questions or queries on these records or any other Vintage records or performers please write to us at Practical Wireless and we'll do our best to help you.

The Mystery of the ST.200

We would, at this point, like to thank the many enthusiasts who have been kind enough to write and help us try to clear up the mystery of the missing John Scott Taggart design "ST.200" mentioned in "Going Back", August 1971.

One theory put forward by **Mr. F. W. Snow**, Belper, Derby, is that there never was a design called the ST.200 but one called "The ST Twin"—a 2-valve reflex. It was published in *The Wireless Constructor* and not the *Popular Wireless* magazine during the late 1924's and early 1925's.

Mr. J. Crossley hailing from Failsworth, Manchester, also maintains that the ST.200 never existed. He

says that J.S.T. visited the United States just after the advent of the ST.100. On his return, he was full of enthusiasm about a gentleman by the name of Armstrong, and *Popular Wireless*-published a circuit by this man which J.S.T. had brought back with him. It was the "Armstrong Regenerative Circuit". J.S.T. then went on to publish the ST.300 design, so it appears that Armstrong's receiver may have taken the place of the missing ST.200.

Mr. A. H. Jenkins, Dringhouses, York, writes to say that as far as he recollects, the ST.200 may well have been just a mains-operated version of the ST.100 and therefore did not differ operationally from that design. It did however, need an extra valve. Mr. Jenkins also informs us that the S.T. series did not start at ST.100, although this was the most famous. He says that he used to have a limp covered book of J.S.T. circuits which started at crystal sets and worked up through one valvers. He seems to remember an ST.37 which was quite well-known at the time.

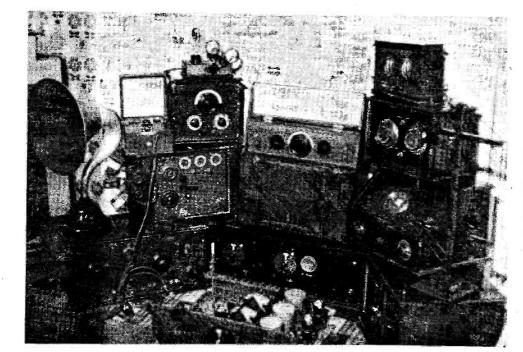
Vintage Sets

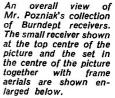
Mr. William Pozniak, from 37 Manley Road, Oldham, Lancashire, has loaned us some photographs. He says that for three years he has been routing through old cellars and derelict radio shops throughout the North (we hope you came out for food now and again, Mr. Pozniak) and has produced the collection of Vintage gear shown in the first picture. He says that it takes him most of his time to strip down and restore these sets to as near as possible their original condition. The sets he has are as follows: The Burndept Screened Four, The Burndept Portable Five, the Burndept Ethophone Five and the Burndept Ethophone Two.

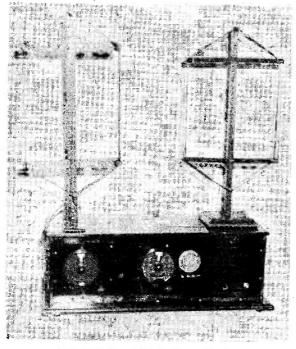
Mr. Pozniak believes there were about 20 Burndept receivers. This figure is agreed by Mr. Keith Lancaster who we mentioned some time ago as wanting to start a Vintage Radio Club. Incidentally, response has not been too good and if readers are interested in such a Club, would they please contact Keith Lancaster, 40 Great Gardens Road, Hornchurch, Essex.

The rest of Mr. Pozniak's collection contains a Marconiphone Type 32 three valver, a domestic Cossor two valver, a small set in a metal case named a GECoPHONE Victor 3 and a few home-constructed receivers.

The receiver shown in the second picture is a



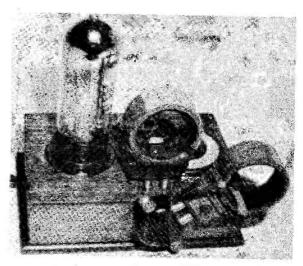




The James Super Six All Waver (c 1927).

1927 James Super Six All Waver (medium, long and short). The reader says that he purchased this set from an old radio shop which was a booming concern in the 1920's, together with a BTH horn speaker, and the person who sold it to him, told him, as he removed the set from the dusty shelf, that he was the first person to touch the set in 40 years.

The neat little receiver shown in the third photograph seems to Mr. Pozniak to be an unusual type, for he says that there appear to be two resistors and



Louwe receiver with encapsulated components.

two capacitors encapsulated in glass tubes which have then been assembled inside the glass envelope of the valve. The make on the case reads "Louwe Radio Berlin" and on the valve it states "Louwe Radio British Made". This set is in working order on headphones although it was originally made to drive some sort of speaker. The reader would like to know the date and origin of this set if anyone can help him and he would also be interested if anyone could supply him with original Burndept valves for any of their makes of receivers.

TITANIC DISASTER

We would like to thank the many readers who have written to us about the Titanic Disaster. We are gradually answering all the letters and we hope you will bear with us.

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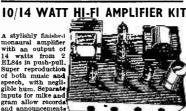
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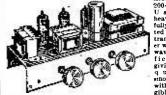
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HIS unit has four switch-selected bands, covering from 1600kHz to 100MHz and it can be used as a wavemeter, field strength meter or phone transmission monitor. It is fully enclosed, selfcontained, and requires no battery.

Figure 1 shows the circuit. One pole of the switch, S1, selects the inductors L1, L2, L3 or L4, which are tuned by VC1. Pole S2 switches the diode D1 to the appropriate small coupling winding. When the circuit is tuned to resonance by VC1, rectified current is indicated by the meter.

To monitor phone signals, high impedance headphones or earpiece can be plugged into the phone jack and the phones are then in series with the meter. Cl is for r.f. and a.f. by-pass purposes, while C2 is an r.f. by-pass capacitor.

INDUCTORS

L1, L2 and L3 are wound on ¹₂in. diameter paxolin formers. These can be prepared by cutting 1^{1}_{2} in. lengths of ¹₂in. diameter tube, and drilling small holes to secure the ends of the windings. The tubes can later be fixed with adhesive, with small brackets or by having them a push fit on small discs of insulating material screwed in place. Surplus or junk box formers could probably be utilised, but if these are of different diameter it may be necessary to modify the number of turns to suit.

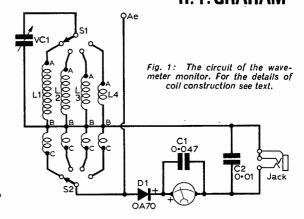
L1 is wound with 34 s.w.g. enamelled copper wire. The tuned portion A to B is 65 turns, in a compact pile about ⁵16in. long. A space of about ²16in. is left and section B to C is wound on in the same direction, and has 8 turns side by side.

L2 also uses 34 s.w.g. enamelled copper wire. A to B is 50 turns wound side by side. A space is left as before, and a 4 turn coupling winding ends at C.

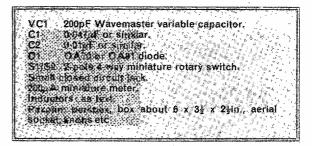
L3 uses 24 s.w.g. enamelled wire and has 12¹₂ turns from A to B, and 2 turns from B to C. Figure 2 shows the general construction of L1, L2 and L3.

The windings are doped, smeared with clear Bostik 1 or with some other suitable coil cement.

L4 is self-supporting, of 16 s.w.g. bare wire, and is shown in Fig. 3. It is wound and shaped to have 8 turns in all and is ¹₂in. outside diameter, and lin. long. B is a tapping 2 turns from end C. This coil is soldered directly to tags A and C of the 4-way switch as shown in Fig. 3. A lead runs from tapping B



★ components list



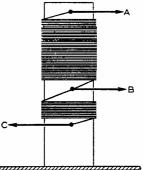


Fig. 2 : The general construction of coils L1, L2 and L3 wound on <u></u>in. diameter paxolin formers. Details of L4 are given in the text and in Fig. 3

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directly to the moving plates tag of the variable capacitor.

CONSTRUCTION

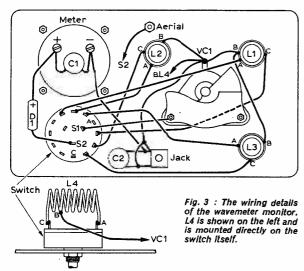
The wavemeter was built to occupy a rigid plastic 'snack box' approximately $6in \times 3^{1}2in \times 2^{1}2in$. deep. A piece of paxolin sheet, a little under $6in \times 3^{1}2in$, was cut to fit inside the box lid. The wavemeter is built wholly on this paxolin panel.

Screws holding L1, L2 and L3 are countersunk. A piece of stout paper to bear dial markings is cut the same size as the paxolin, and cemeted to it. The wavemeter is then assembled and calibrated and the scales are marked. The meter and fixing nuts of the switch, jack, and VC1 are then removed and the paxolin panel is fitted inside the box lid. This places the scales under the transparent material of the lid. Paxolin and lid are held together by the meter and the nuts mentioned.

Holes for the jack, switch and VCl are first drilled in the paxolin and the meter hole cut with an adjustable washer or tank cutter. The paxolin is then placed in the lid, and the positions of holes marked on the lid itself, so that they can be made to match up correctly with those in the paxolin.

The box should be of material which is reasonably strong. It should be drilled and cut with tools which are really sharp, and without forcing, to avoid cracking.

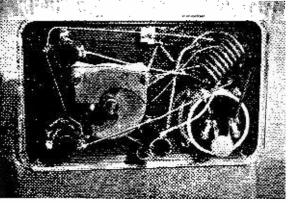
A suitable case could be assembled from ${}^{1}_{4}$ in. thick wood, glued and pinned at the joints. The paxolin panel would then be large enough to be fixed with small wood screws.



All wiring is shown in Fig. 3. Connections should be of stout wire and run clear of each other, and should be reasonably short and direct, especially for L3 and L4.

VCl is 200pF, but this exact value is not essential, provided that the ranges overlap a little. If an alternative diode is fitted, be sure it is suitable for frequencies up to 100MHz.

A 100μ A meter would be a little more sensitive than the one specified, but would also cost a little more. Should a 500μ A meter be to hand, it is suitable, though less sensitive. The sensitivity with a 1mA meter becomes rather low so an instrument in the neighbourhood of 200μ A is probably most suitable.



An internal view of the completed prototype.

The dial cursor is a 2^{3}_{4} in. piece of perspex about ${}^{5}_{2}$ in. wide, marked with a hair line. Readings will be confused if VC1 can rotate through 360°, so the extreme corner of the back moving plate was turned up with pliers to limit rotation to about 180°.

Values used for C1 and C2 are not particularly important. A medium or high impedance pair of headphones will give best general results.

CALIBRATION

Actual band coverage with the inductors described is as follows

Range 1	1∙6-4MHz
Range 2	2.8-10MHz
Range 3	8-32MHz
Range 4	21-100MHz

This gives 160 and 80m bands in Range 1, 80 and 40m bands in Range 2, all h.f. bands 20-10m in Range 3, and 15-10m bands and harmonics in Range 4, without having to use the extreme h.f. end of the ranges, where readings become close together.

The simplest means of calibration is to use a grid dip oscillator, placing its coil 2in. or so from the appropriate inductor in the wavemeter. This allows complete calibration of all ranges with a minimum of difficulty.

A transmitter can be used for calibration in the amateur bands, and these will probably be of most importance. Place the wavemeter some distance from the transmitter, its dummy load or the aerial lead or tuner, as required to obtain a deflection on the meter.

Some signal generators have an r.f. level meter. If the output of the generator is coupled to the wave meter by a small loop, a dip may be observed on the level meter when the wavemeter is tuned to resonance. This offers another possible way to secure calibration.

If the wavemeter is coupled to a receiver having a sensitive S-meter, a *slight* dip in the S-meter reading can be observed when the wavemeter is tuned to the receiver frequency. Input to the receiver can be from a signal generator, or transmission.

All calibration is made with no external connection to the wavemeter and minimum coupling to avoid unnecessary inaccuracy.

USES

A wavemeter is normally used to indicate frequency bands and harmonics, not exact frequencies within a narrow band. It is placed or held to pick up sufficient r.f. energy from the circuit being investiga-

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	TRANSISTORS	······································		C. NEW PRICES
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Na004 460p 28501 322p BC. N3055 62p 28502 35p BC. N3134 30p 28503 271p BC. N3134 30p 3808 40p BC. N3134 30p 38128 40p BC. N3135 25p 3N142 70p BC. N3136 25p 3N147 721p BC. N3300 25p 3N140 771p BC. N3301 25p 3N142 55p BC. N3301 20p 3N142 55p BC. N3300 250p 3N142 55p BC.	1860 624p BFV90 674p NKT24 1867 119 BFW58 271p NKT24 1888 10p BFW59 25p NKT24 1888 11p BFW60 25p NKT24 1898 11p BFW25 61.86 NKT24 1896 129 BFX29 61.80 NKT24 130 124p BFY10 61.84 NKT26 171 159 BFX39 474p NKT26	0 2740 T1860 224p 1 274p T1861 25p 2 20p T1862 274p 3 624p T1P29A 50p 4 174p T1P30A 60p 5 20p T1P31A 624p 1 20p T1P32A 75p 2 30p T1P32A	HEAT SINKS 4·8 × 4 × lin Finned for Two TO-3 Trans., 474p. 4·8 × 2 × lin Finned, for One TO-3 Trans., 824p. For SO-1, 84p. For TO-5. 5p Finned. For TO-18, 5p Finned. For TO-18, 1/- Finned.	Log. and Lin., less switch, 16p. Log. and Lin., with switch, 25p. Wire-wound Pots (20%), 40p. Twin Ganged Stereo Pots, Log and Lin., 40p. PRESETS (CARBON) 0-1 Watt 6p VERTICAL
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A series of simple transistor projects, each using less than twenty components and costing less than one pound to build.

"W HAT's an automatic light?" you may well ask. Well, lights are only needed when it's dark, and a very simple electronic switch circuit can be made to turn a small light on as soon as the general light level drops below a certain level; this has its most obvious application in motor cars. Many parts of the country have local regulations requiring cars to display lights shortly after sunset, even when parked. In other areas car parking lights are needed only when the local street lighting is turned off, which in some areas is as late as 2 a.m. There cannot be many people who relish the thought of going out to their cars at that hour to turn on the side or parking lights.

Porch lights which automatically turn themselves on after dusk are another application and although the circuit as shown is not really meant for this, many readers will realise the potential and modify the circuit to operate a thyristor or relay to apply the mains supply to the porch light.

THE CIRCUIT

Like last month we are making use of a light dependent resistor, that is a resistor whose value in ohms falls with an increase in light falling on it. Also like last month this component is coupled into the base bias potential divider circuit, thus controlling the bias on the transistor, Trl. R1 may have to be altered in value in order to make the "switch" operate at the correct light level.

As darkness falls, the potential at the base of Tr1 falls as the value of the LDR rises and after a certain point Tr1 begins to turn off, less current is passed and the potential between the emitter and the collector rises. This in turn starts to turn Tr2 on. This transistor, when light was falling on the LDR, thus holding Tr1 on, was off since the potential between base and emitter was held at a very low level due to the small potential across Tr1. As soon as Tr2 begins to conduct this increases the current through R4, thus causing Tr1 to cut off even more and in turn switching Tr2 more on. This action takes place very quickly and Tr2 is switched fully on, applying the supply voltage across the bulb causing it to light up.

This switching circuit is known as the Schmitt Trigger and a very useful circuit, it is too, with applications in many fields of electronics. As can be seen, only a few components are required and the switching action, using modern transistors, is very rapid and very positive. As soon as the potential at the base of Trl either reaches or falls to a preset level, Tr2 will switch.

The supply necessary to operate the circuit is taken directly from the car battery, though it is good practice to take this via the car's fuse as any

-continued overleaf

No. 30 Automatic light

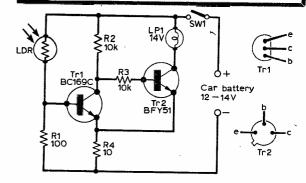


Fig. 1: The circuit of the automatic light with the lead connections of the transistors shown on the right.

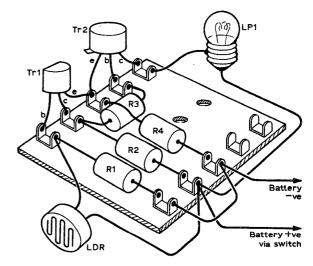
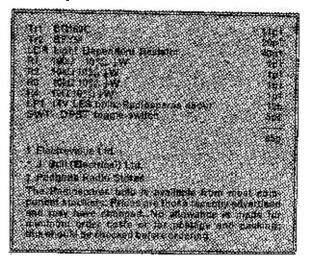


Fig. 2: A suggested component layout on a small piece of tag board.

\star components list



TAKE 20—continued from previous page

mistakes or serious component failure could create a fire risk.

Car batteries are capable of giving very high currents and therefore the quiescent current is of no great importance but it is in fact low, in the order of 2mA. The recommended bulb draws 40mA and it can be seen that these currents can be supplied by a number of dry batteries, though operation in this manner will not be cheap.

The bulb should last a considerable time as it is being slightly under-run; there is of course a small volts drop across R4 and Tr2.

CONSTRUCTION

A suggested layout is shown in Fig. 2 but this is not critical. One point must be watched however; if the light from the bulb is allowed to fall onto the LDR it will switch the circuit in the same way that sunlight will. This will cause the circuit to oscillate; this can easily be seen as the bulb will flicker or flash. To avoid this the LDR must be pointed away from the bulb.

WAVEMETER MONITOR-by R. F. GRAHAM

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ted. Tuning resonance is shown by maximum meter reading. In driver and similar circuits with small power, the wavemeter needs to be reasonably close to the circuit inductor. For P.A. and high power circuits, it can be at some small distance.

To check the frequency of low level multipliers and similar circuits, or those which are inaccessible, a link line may be used. This is twin flex or co-axial cable about 2ft. long with a loop of two or three turns each end. One loop is held near the equipment inductor, and the other near the wavemeter inductor. Clipping a lead to jack and aerial socket also allows the coil primaries to be used for coupling. Resonance will be shown by a dip in grid current in a subsequent stage.

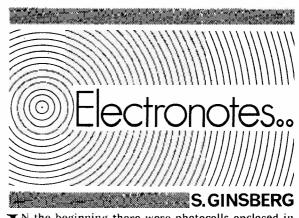
To monitor the audio of an outgoing transmission, plug in phones, tune to resonance and situate the unit where pick-up gives adequate volume.

To tune an aerial, transmitter, tuner or coupling device for maximum radiation from the aerial, a vertical rod or wire is plugged into the aerial socket of the unit, which should be at some distance. Adjustments are then for maximum meter reading, which corresponds to maximum field strength. It is also possible to link couple a dipole or aerial-earth system to the wavemeter for this purpose, or take a short aerial to VC1, ignoring calibration.

The wavemeter produces no signal. It can be used to find bands on a t.r.f. receiver by adjusting the latter until oscillation *just* arises, and holding the wavemeter coil in line with the receiver coil. Resonance when tuning the wavemeter is then shown by the receiver going out of oscillation.

BLUEPRINT SERVICE

We would like to draw readers' attention to the fact that the BLUEPRINT SERVICE has been discontinued and therefore no further BLUEPRINTS are available.



I N the beginning there were photocells enclosed in glass envelopes. Variations of these were more sensitive and multi-electrode secondary emission cells appeared like the 931A photomultiplier tube.

In more recent years, many experimenters, and indeed, most of the electronics industry too, has opted for solid state when thinking in terms of a light sensing device. An example here is the ORP12 light-dependent resistor, not to mention the earlier plastic-encapsulated transistors with some of their paint removed!

In America, work is currently proceeding on a technique known as charge coupling for imaging devices. One oxidised surface of a tiny silicon chip is covered with some 288 minute electrodes. These are arranged into groups of three, the last electrode in each group being a common conductor for the whole array.

The area immediately beneath each group of three elements forms a tiny light sensitive device in its own right. Minority carriers are generated in the silicon beneath the electrodes whenever light is focussed onto them. The greater the intensity of the light, the greater the charge, thus forming a direct relationship between light intensity and electrical charge.

The centre electrode of each group is purposely made more positive than the other two elements and thus the charges on minority carriers are attracted to, and congregate at, these centre electrodes.

Individual charges are next moved along the array from one electrode group to the next by putting a successively greater positive voltage on the electrode next to the one which is holding the charge. Eventually all the charges will have been passed along to the end of the device where they are collected by an output electrode.

Because the size of the charge will be dependent upon the amount of original light applied, the final electrode has, in effect, an analogue electrical signal which has a linear relationship to original light patterns of intensity.

How big is a $100,000\mu$ F capacitor—think of the size of some electrolytics. Now imagine how big a one Farad capacitor would be. Just to stretch the imagination even further, think of a 50 Farad capacitor. Fantastic—but they exist and are for sale. Twist in the tale is that the 50F is contained within one-third of a cubic inch. The devices will also retain their storage (if charged with a voltage) and leakage is so low that even after 16 months they will have retained over 97 per cent of the original charge. Think; they could be charged up and used as a form of battery. When discharged, they could simply be recharged again ad infinitum.



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117

2.4

(AA)

0.14 000

1118

SEMI-CONDUCTORS/VALVES

Transistors 2N3415 22p 2N5458 2N3416 27p 2N5459	85p BC114 15p BFW90 22p NKT219 80p 48p BC115 15p BFW91 20p NKT223 27p	integrated FJH101 25p 8N7430 23p Circuits FJH111 70p 8N7440 23p	VALVES
2G301 20 p 2N3417 37 p 28102 2G302 19 p 2N8439 130 p 28103 2G303 20 p 2N3440 97 p 28104	25p BC116 15p BFX12 22p NKT224 22p 25p BC118 15p BFX13 22p NKT225 22p 25p BC119 35p BFX29 25p NKT229 30p	FJH121 25p SN7441AN CA3000 180p FJH131 25p 87p CA3005 117p FJH141 25p SN7442 87p	OA2 38p 25Z4 80p EL95 35p OB2 45p 25Z5 42p EM80 40p OZ4 30p 25Z6 55p EM81 50p
2G306 80p 2N3564 17p 2S301 2G308 80p 2N3565 15p 2S302 2G309 80p 2N3566 22p 2S303	47p BC121 20p BFX30 25p NKT237 35p 47p BC122 20p BFX37 30p NKT238 25p 57p BC125 15p BFX44 37p NKT240 27p	CA3007 262p FJH151 25p 8N7446 125p CA3011 75p FJH161 70p 8N7447 110p CA3012 88p FJH171 25p 8N7448 100p	IL4 20p 30C15 60p EM84 85p IR5 35p 30C17 85p EM85 21.00
2G371 15p 2N3568 25p 28304 2G374 20n 2N3569 25n 28501	720 BC126 250 BFX68 670 NKT241 270 820 BC134 150 BFX84 250 NKT242 200	CA3013 105p FJH181 25p 8N7450 28p CA3014 124p FJH221 25p 8N7451 28p	IT4 25p 30F5 85p EY51 40p IU4 27p 30FL1 70p EY86 40p
2N388A 49p 2N3572 97p 28503 2N404 20p 2N3605 27p 3N83	279 BC136 15p BFX86 25p NKT244 17p 40p BC137 15p BFX87 25p NKT245 80p	CA3018A FJH241 25p SN7454 23p 110p FJH251 25p SN7460 23p	IU5 509 30FL12 929 EY87 429 2D21 359 30FL14 759 EZ40 459 3Q4 409 30L15 859 EZ41 459
2N696 15 p 2N3606 27 p 3N128 2N697 15 p 2N3607 22 p 3N140 2N698 25 p 2N3638 18 p 3N141	70p BC138 37p BFX88 20p NKT261 20p 77p BC140 35p BFX89 82p NKT262 30p 72p BC141 35p BFX93A 70p NKT264 20p	CA3019 84p FJJ101 50p 8N7472 85p CA3020 126p FJJ111 50p 8N7473 43p CA3020A FJJ121 60p 8N7474 43p	384. 359 30L17 809 EZ80 259 3V4 459 30P12 809 EZ81 289 5R4 609 30P19 809 GZ32 489
2N699 429 2N3638A 209 3N142 2N706 99 2N3641 189 3N143 2N706A 119 2N3642 189 3N152	559 BC147 159 BFY11 429 NKT271 209 679 BC148 119 BFY18 259 NKT262 209 879 BC149 159 BFY19 259 NKT274 209	160p FJJ131 60p SN7475 47p CA3021 156p FJJ141 125p SN7476 45p CA3022 130p FJJ181 75p SN7483 87p	5U4 880 30PL1 700 GZ34 600 5V4 420 30PL13 930 KT66 42:05 5Y3 820 30PL14 900 KT88 42:00
2N708 150 2N3643 200 40050 2N709 450 2N3644 250 40250 2N718 250 2N3645 250 40251	50p BC152 17p BFY21 42p NKT275 20p 50p BC153 35p BFY24 45p NKT278 25p 82p BC154 35p BFY29 40p NKT281 27p	CA3023 1265 FJJ191 655 8N7486 505 CA3026 1005 FJJ211 1255 8N7490 100 p CA3028A 745 FJJ251 1255 8N7493 875 CA3028A 745 FJJ251 1255 8N7493 875	5Z4G 40p 35L6 50p MU14 60p 8/30L2 75p 35W4 80p PABC80 40p 6AC7 25p 35Z4 80p PC86 60p
2N718A 30p 2N8691 15p 40309 2N726 25p 2N8692 18p 40310 2N727 25p 2N8693 15p 40311	829 BC157 159 BFY30 409 NKT401 879 459 BC158 159 BFY41 509 NKT402 909 859 BC159 159 BFY43 629 MKT403 756	CA3028B FJL101 125p SN7493 87p 105p FJY101 25p SN7495 87p CA3029 87p IC10 250p SN7496 87p	6AG7 400 35Z5 400 PC88 600 6AK5 800 50B5 450 PC97 450 6AK6 570 50C5 400 PC900 430
2N914 17 p 2N3694 18 p 40312 2N916 17 p 2N3702 10 p 40314 2N918 30p 2N3703 10 p 40315	47p BC160 35p BFY50 22p NKT404 62p 37p BC167 15p BFY51 20, NKT405 75p 37n BC168B 14p BFY52 22p NKT406 62n	CA3029A IC12 250p SN74107 43p 165p L900 40p SN74153 CA3030 137p L914 40p 140p	6AL5 20p 80 50p PCC84 40p 6AM6 33p 85A2 45p PCC85 40p 6AQ5 35p 807 50p PCC83 55p
2N929 229 2N3704 129 40316 2N930 249 2N3705 109 40317 2N937 529 2N3706 109 40319	47p BC168C 15p BFY58 17p NKT451 62p 37p BC169B 14p BFY56A 57p NKT452 62p 55p BC169C 15p BFY56A 57p NKT453 47p	CA3035 122p L923 40p SN74154 CA3036 72p MC724P 66p 220p CA3039 82p MC780P 247p SN74160	6AS6 879 1025 509 POC89 509 6AT6 309 5763 709 POC189 559 6AU6 259 6146 \$1.50 POF80 309
2N1090 220 2N3707 120 40320 2N1091 220 2N3708 80 40323 2N1131 250 2N3709 100 40324	479 BC170 129 BFY77 579 NKT713 209 829 BC171 159 BFY90 659 NKT717 429 479 BC172 159 BSX19 179 NKT734 275	CA3041 109p MC788P 82p 180p CA3042 109p MC790P 124p SN74161 CA3043 137p MC792P 66p 180p	6AV6 80p AZ81 50p PCF82 34p 6BA6 25p CY31 35p PCF84 50p 6BE6 30p DAF91 25p PCF86 60p
2N1132 25p 2N3710 10p 40326 2N1302 17p 2N3711 10p 40329 2N1303 17p 2N3713 187p 40344	879 BC175 229 BSX20 179 NKT736 859 809 BC177 229 BSX21 209 NKT773 259 279 BC178 209 BSX26 459 NKT781 809	CA3044 1209 MC799P 669 SN74164 CA3045 1229 MC1303L 2209 CA3046 819 2629 SN74165	6BH6 45p DAF96 42p PCF800 80p 6BJ6 45p DF91 25p PCF801 50p 6BQ7A 40p DF96 42p PCF802 50p
2N1304 22p 2N3714 200p 40347 2N1305 22p 2N3715 222p 40348 2N1306 24p 2N3716 290p 40360	577 BC179 229 BSX27 479 OC16 509 529 BC182 129 BSX28 329 OC19 379 429 BC182L 109 BSX60 329 OC20 759	CA3047 1879 MC1304P CA3048 2049 8759 8N74192 CA3049 1609 MC1305P CA3050 1859 8869 8N74193	6BR7 85p DK91 35p PCF805 75p 6BR8 65p DK92 50p PCF806 70p 6BW6 85p DK96 42p PCF808 75p
2N1307 24p 2N3773 240p 40361 2N1308 25p 2N3791 275p 40362 2N1309 24p 2N3819 34p 40370	47p BC183 10p B8X61 62p OC22 50p 55p BC183L 10p B8X76 15p OC23 60p 82p BC184 15p B8X77 20p OC24 60p	CA3050 185p 386p 8N74193 CA3051 134p MC838P 285p CA3052 165p 549p TAA241	6BW7 70p DL92 35p PCL82 35p 6BZ6 35p DL94 45p PCL82 65p 6C4 33p DL96 42p PCL84 45p
2N1607 17p 2N3820 57p 40406 2N1613 21p 2N3823 75p 40407 2N1631 85p 2N3854 87p 40408	57p BC184L 12p BSX78 25p OC26 87p 40p BC186 25p BSY24 15p OC26 25p 52p BC187 27p BSY25 15p OC28 60p	CA3053 46p MC1435P 162p CA3054 109p 345p TAA242 CA3055 240p MC1552G 425p	6CD6 51:15 DM70 38p PCL85 40p 6CL6 50p DY86 33p PCL85 40p 6CW4 63p DY87 35p PFL200 70p
2N1632 30p 2N3854A 27p 40409 2N1637 30p 2N3855A 27p 40410 2N1638 27p 2N3855A 30p 40412	559 BC212L 129 BSY26 179 OC29 60p 629 BC213L 129 BSY27 179 OC35 509 509 BC214L 159 BSY28 179 OC36 60p	CA3059 165p 461p TAA243 150p CA3064 180p MC1709CG TAA263 75p PCH101 85p 94p TAA293 97p PCH101 105p MFC4000P TAA200 175p	6F1 62p E88CC 65p PL36 55p 6F6G 30p E180F 95p PL81 50p
2N1639 27p 2N3856 30p 40467A 2N1701 110p 2N3856A 35p 40468A	57p BCY10 27p BSY29 17p OC41 22p 35p BCY30 24p BSY32 25p OC42 25p	FCH111 105p MFC4000P TAA300 175p FCH121 105p 112p TAA310 125p FCH131 50p PA222 487p TAA320 72p	6F14 65p EAF42 85p PL83 45p 6F15 65p EB91 20p PL84 40p
2N1889 32p 2N3858A 30p 40600 2N1893 37p 2N3859 27p 40603	50p BCY32 50p BSY37 25p OC45 12p 50p BCY33 20p BSY38 20p OC46 15p	FCH141 105p PA230 100p TAA350 175p FCH151 105p PA234 100p TAA350 175p FCH161 50p PA234 100p TAA435 147p FCH161 50p PA237 185p TAA521 132p	6F18 45p EBC41 55p PL500 75p 6F23 80p EBC81 80p PL504 80p 6H6 20p EBF80 40p PY32 55p
2N2160 57p 2N3860 30p AC126 2N2193 40p 2N3866 150p AC127	30p BCY34 25p BSY39 22p OC70 15p 20p BCY38 80p BSY43 50p OC71 12p 24p BCY39 80p BSY41 82p OC72 12p 20p BCY30 80p BSY51 82p OC72 12p 20p BCY40 50p BSY52 32p OC73 30p	FCH171 105p PA246 945p TAA522 860p FCH181 105p PA424 285p TAA530 495p FCH191 105p PA424 447p TAA811 445p	6J4 500 EBF83 400 PY33 680 6J5 200 EBF89 820 PY80 850 6J5GT 300 EBL21 600 PY81 309 6J6
2N2193A 42p 2N3877 40p AC128 2N2194 27p 2N3877A 40p AC151 2N2194A 30p 2N3900 37p AC152 2N2217 37p 2N3900 40p AC154	Boy BCY41 J5p BSY53 S7p OC74 S0p 22p BCY42 15p BSY54 40p OC76 22p 22p BCY43 15p BSY56 90p OC76 22p	FCH201 180p PA265 497p TAB101 97p FCH201 180p SN7400 23p TAD100 150p FCH221 180p SN7401 28p TAD100 197p	6J6 20p EC86 60p PY82 30p 6J7 45p EC88 60p PY83 38p 6K8G 35p ECC40 60p PY83 40p
2N2217 275 2N3904 405 AC176 2N2218 205 2N3901 975 AC176 2N2219 205 2N3903 255 AC187 2N2220 255 2N3904 255 AC188	25p BCY54 25p BSY79 45p OC77 20p 25p BCY58 22p BSY90 57p OC78 20p	FCH231 150p SN7402 23p SL403A 187p FCJ101 160p SN7403 23p SL702C 147p FCJ111 150p SN7404 23p UA702A 280p	6L6GT 45p ECC84 80p PY800 50p 6LD20 40p ECC85 60p PY801 50p 6Q7 40p ECC88 40p U25 75p
2N2222 25p 2N3905 30p ACY17 2N2222 20p 2N3906 30p ACY17 2N2222 20p 2N3906 30p ACY18 2N2222A 25p 2N4058 15p ACY19	27p BCY59 22p BSY95A 12p OC81 20p 27p BCY60 97p C424 15p OC81D 20p 24p BCY70 15p OC82 25p . 25p . 24p BCY70 15p OC82 25p . <td< td=""><td>FCJ121 275p SN7405 23p UA702C 77p FCJ131 275p SN7406 80p UA703C 187p FCJ131 525p SN7408 83p UA703C 187p FCJ141 525p SN7408 23p UA709C 125p</td><td>68A7 409 ECF80 859 U26 759 68G7 359 ECF82 359 U50 329 68J7 409 ECF86 659 U52 339</td></td<>	FCJ121 275p SN7405 23p UA702C 77p FCJ131 275p SN7406 80p UA703C 187p FCJ131 525p SN7408 83p UA703C 187p FCJ141 525p SN7408 23p UA709C 125p	68A7 409 ECF80 859 U26 759 68G7 359 ECF82 359 U50 329 68J7 409 ECF86 659 U52 339
2N2227 30p 2N40059 10p ACY20 2N2368 15p 2N4060 12p ACY20 2N2368 15p 2N4060 12p ACY21 2N2369 17p 2N4061 12p ACY22	SOP BCY72 15p GET113 20p OC83 25p SOP BCY72 15p GET113 20p OC83 25p SOP BCY78 30p GET114 15p OC84 25p 10p BCY79 30p GET118 20p OC139 25p	FCJ201 100p SN7409 23p UA710C 125p FCJ211 975p SN7410 23p UA716 187p	68K7 859 ECH21 879 U191 759 68L7 859 ECH21 879 U191 759 68L7 859 ECH22 709 U281 409 68N7 859 ECH42 709 U282 409 68Q7 409 ECH41 809 U301 409
2N2369A 17p 2N4062 12p ACY28 2N2410 42p 2N4244 47p ACY39 2N2483 87p 2N4248 15p ACY40	17p BCZ10 27p GET120 25p OC140 38p 47p BCZ11 40p GET873 12p OC170 25p 14p BD112 50p GET880 30p OC171 30p	FCK101 430p SN7411 22p UA723C 182p FCL101 230p SN7413 35p UA733C 160p FCU101 230p SN7420 23p UA733C 182p FCU101 102p SN7420 23p UA741C 87p	6SQ7 40p ECH81 30p U301 40p 6U4 60p ECH83 40p U801 \$1-00 6V6G \$5p ECL80 40p UABC80 \$5p 6V6GT \$2p ECL82 \$5p UABC80 \$5p
2N2484 82p 2N4249 15p ACY41 2N2539 22p 2N4250 18p ACY44 2N2540 22p 2N4254 42p AD140	250 BD116 1120 GET887 150 OC200 400 400 BD121 650 GET889 220 OC201 500 470 BD123 800 GET890 220 OC202 750	BRIDGE 50 PIV 4A 60p RECTIPIERS 100 PIV 4A 70p PLASTIC 200 PIV 4A 75p	6X4 80p ECL88 65p UBC41 50p 6X5G 80p ECL86 40p UBC81 40p 6X5GT 27p EF37A 60p UBF80 40p
2N2613 27p 2N4255 42p AD149 2N2614 30p 2M4284 17p AD150 2N2646 47p 2N4285 17p AD161	479 BD124 759 GET896 829 OC203 409 629 BD131 759 GET897 229 OC204 409 359 BD132 859 GET898 829 OC205 759	ENCAPSULATED 400 PIV 4A 80p 600 PIV 1A 50p 50 PIV 6A 62p 50 PIV 2A 55p 100 PIV 6A 75p	10C2 509 EF39 409 UBF89 359 10F1 909 EF40 509 UCC84 499 10F13 559 EF41 659 UCC85 409
2N2711 25p 2N4286 17p AD162 2N2712 25p 2N4287 17p AF109 2N2713 27p 2N4288 15p AF114	85p BDY10 125p MAT100 25p OC206 90p 45p BDY20 105p MAT101 30p OC207 75p 85p BDY61 125p MAT102 25p OCP71 42p	100 PIV 2A 609 200 PIV 6A 859 200 PIV 2A 679 400 PIV 6A \$1.10 400 PIV 2A 809	10P14 41-10 EF42 70p UCF80 55p 12AT6 30p EF80 25p UCH21 60p 12AT7 30p EF85 35p UCH42 70p
2N2714 80p 2N4289 15p AF115 2N2904 20p 2N4290 15p AF116 2N2904A 25p 2N4291 15p AF117	25p BDY62 100p MAT121 80p ORP12 50p 25p BF115 25p MJ400 107p ORP60 40p 20p BF117 47p MJ420 112p ORP61 42p	SILICON RECTIFIERS MINIATURE WIRE ENDED PLASTIC IN PL CL	12AU7 300 EF80 300 UCH81 35p 12AU7 300 EF89 380 UCH81 35p 12AV7 300 EF89 389 UCL82 359 12AV6 33p EF91 33p UCL83 60p
2N2905 25p 2N4292 15p AF118 2N2905A 30p 2N4294 17p AF121 2N2906 90p 2N4803 47p AF124	44p BF152 28p MJ421 112p P346A 22p 30p BF154 38p MJ430 102p ST140 15p 92p BF158 28p MJ440 95p ST141 90p	SERIES SERIES SERIES 1 AMP 1.5 AMP 3 AMP 4001 50PIV 08p 10p 19p	12BA6 35p EF92 40p UF41 60p 12BE6 35p EF183 30p UF80 35p 12BH7 40p EF184 35p UF85 40p
2N2906A 25p 2N4964 15p AF125 2N2907 23p 2N4965 18p AF126 2N2923 15p 2N5027 52p AF127	19p BF159 57p MJ480 97p TI834 62p 16p BF163 85p MJ481 125p TI843 40p 16p BF167 95p MJ490 100p TI844 19p	4002 100PIV 09p 11p 20p 4003 200PIV 10p 12p 22p 4004 400PIV 10p 12p 25p	19AQ5 85p EH90 40p UF89 85p 20D1 45p EL34 50p UL41 65p 20F2 75p EL33 \$1.25 UL84 \$0p
2N2924 15p 2N5028 57p AF139 2N2925 15n 2N5029 47p AF178	25p BF170 83p MJ491 187p T1845 12p 42p BF177 80p MJE340 62p T1846 12p	4005 600PIV 12p 18p 28p 4006 800PIV 15p 17p 27p 4007 1000PIV 20p 20p 30p	20L1 £1-10 EL41 55 p UY41 45 p 20P1 50 p EL42 58 p UY85 30 p
2N2926O 12p 2N5172 12p AF180 2N2926Y 12p 2N5174 52p AF181 2N3011 24p 2N5175 52p AF186	400 BF178 200 MJE371 870 TIS49 120 400 BF179 800 MJE520 870 TIS49 120 880 BF180 850 MJE531 870 TIS49 170	50+ less 15% 100+ less 20%	20P3 60p B1.81 55p VR105/30 88p 20P4 £1.10 EL84 25p VR150/30 35p. 20P5 £1.40 EL85 43p Add 12p in \$ 20F5 £1.90 EL85 43p Add 12p in \$
2N3014 25p 2N5176 45p AF239 2N3053 20p 2N5232A 80p AF279 2N3054 49p 2N5245 45p AF280	47p BF182 80p MPF103 80p T1802 12p 47n BF184 90n MPF104 87n T1853 92n	STUD MOUNTING 6A 10A 17.5A 35A 100PIV — 45p 50p \$1.28	DIODES & RECTIFIERS
2N3055 72p 2N5246 42p AFZ11 2N3133 25p 2N5249 67 ASY26 2N3134 80p 2N5265 325p ASY27	825 BF185 205 MPF105 375 XB112 125 255 BF194 175 MP83638 329 XC141 355 315 BF195 156 NKT124 428 ZTX107 145	200PIV 25p 50p 55p £1.42 400PIV 30p 55p 65p £1.42 600PIV 32p 60p 72p £1.13 800PIV 32p 50p 57p £24 £2	IN34A 10p BA154 12p GJ7M 87p 1N914 7p BAX13 12p OA5 17p 1N916 7p BAX16 7p OA6 12p AA119 7p BAX31 7p OA10 22p
2N3135 25p 2N5305 37p ASY28 2N3136 25p 2N5306 40p ASY29 2N3390 25p 2N5307 37p ASY50	24p BF196 15p NKT125 27p ZTX108 11p 27p BF197 15p NKT126 27p ZTX109 17p 25p BF198 42p NKT128 27p ZTX109 11p	800PIV 355 759 875 43 47 1000PIV 40p 859 \$1.65 \$2.77 50 + less 15% 100 + less 20%	AA129 10p BAY38 17p OA9 10p AA213 10p BY100 15p OA47 7p AA215 12p BY103 22p OA70 7p
2N3391 209 2N5308 379 ASY51 2N3391A 309 2N5309 529 ASY54 2N3392 179 2N5310 429 ASY67	829 BF200 359 NKT135 279 ZTX301 179 259 BF224 209 NKT137 329 ZTX302 189 459 BF225 900 NKT210 300 ZTX303 189	ZENER DIODES 400MW 1.5 WATT 10 WATT 3.3-33 V 2.4-100 3.9-100V	BA100 159 BY122 879 0A73 109 BA102 239 BY124 159 0A79 99 BA110 239 BY126 159 0A81 79 BA110 259 BY126 159 0A81 79
2N3393 15p 2N5354 27p ASY86 2N3394 15p 2N5355 27p ASZ21 2N3402 22p 2N5356 32p AUY10	82p BF287 23p NKT211 80p ZTX804 25p 87p BF288 82p NKT212 80p ZTX500 16p	15p each 20p each 25p each 25+ less 15% 100+ less 20%	BA111 87p BY127 17p OA85 7p BA112 70p BY164 57p OA90 7p BA115 7p BY210 85p OA91 7p
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EBF89	12±p	PCF82	121p	6U4	10p
ECC81	10p	PL36	20p	6F23	20 p
ECC82	12½p	PL81	17 <u>∔</u> p	20P1	20p
ECC83	12 ¹ / ₂ p	PY81	7±p	20P3	10p
ECL80	7 <u>∓</u> p 7 <u>∓</u> p	PY33	17±p	20D1	10p
EF80	7 <u></u> ±p	PY82	7±p	30P4	20p
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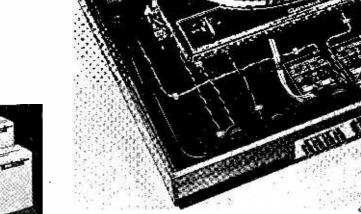
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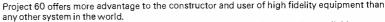
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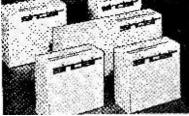


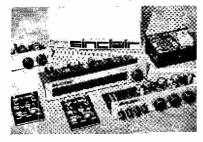
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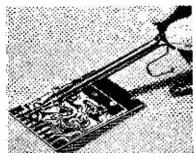
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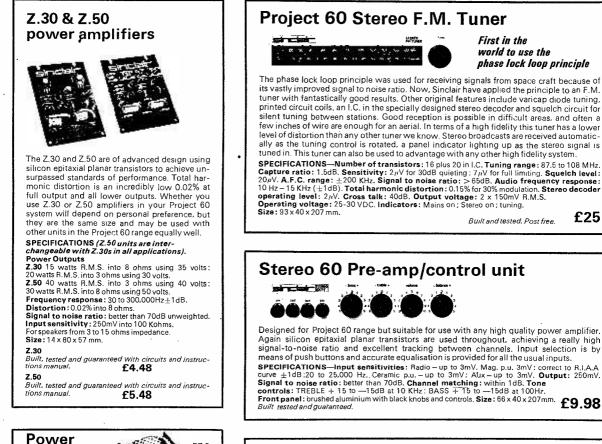




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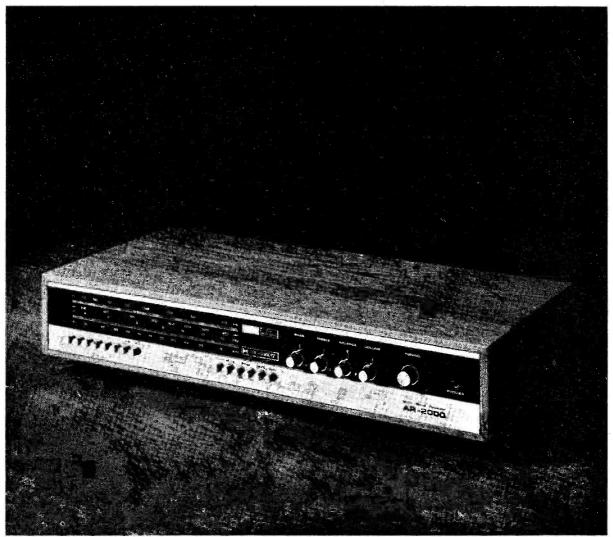
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AKAI AA 3550 AKAI AA 3550 AKAI 6000. ARENA 2000 ARENA 2	51-265 45-50 th MPX 229-00 142-53 128-850 303-45- 303-45- 99-50 91-59-91-50 303-45- 91-59-91-50 303-45- 91-59-91-50 103-45 139-00 160-00 194-74 111-10 110-00 51-50 155-00 149-00 catridge- magneti nplifiers iseption - isted. 13-00 18-80 18-80 18-80 18-80 18-80 120-00 18-80 120-00 100-00 10	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	



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EMPIRE 999/X		11-50	9 25	
EMPIRE 909E/X		12.85	10.25	
EMPIRE 909/X		8·00	7.50	
EMPIRE 90EE/X		9.75	8.00	
ORBIT Magnetic NM 22		cial Pri		1
PICKERING V15 AC2		8.40	7.00	
ORTOFON SL15E		29.65	23.75	
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SHURE M32E		11.10	8-25	
SHURE M32-3		10.20	7.95	
SHURE M44-5	••••		8-00	
SHURE M44-7		10.20	8-00	1
SHURE M-44C		12.05	8-25	1
SHURE M44E		12.95	9 50	
SHURE M55E		17.60	14-00	
		16.70	13.00	
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SHURE M75E-95G		23.15	18.00	
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watts RMS, and 2 KELETRON KN500 Speakers, each speaker		
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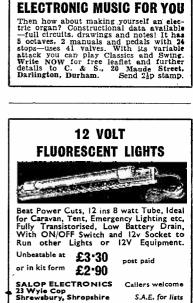
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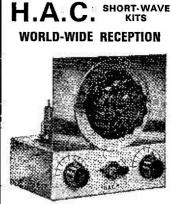


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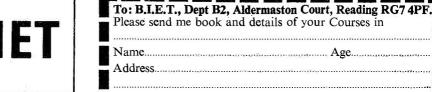
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