

## ADCOLA Soldering Instruments add to your efficiency

- THE NEW 'INVADER'


## ADCOLA L. 646

for Factory Bench
Line Assembly
A precisioninstrument-supplied with standard $3 / 16^{\prime \prime}(4.75 \mathrm{~mm})$ diameter, detachable copper chisel-face bit*.
Standard temp. 360 c at 23 watts.
Special temps. from $250^{\circ} \mathrm{c}$ 410 c.

## *Additional Stock Bits

(illustrated) available
COPPER

| - | , |
| :---: | :---: |
| B 38 : -3.2 mm | chisel fac |
| E--- | $\rightarrow$ |
| B $143-2.4 \mathrm{~mm}$ | chisel |
| B24 | T |
| B $24 \therefore-4.75 \mathrm{~mm}$ | ctick |
|  |  |
| B $12 \quad 2.4 .75 \mathrm{~mm}$ | eriet bit |
| ---- |  |

LONG LIFE


Don't take chances. We don't. All our ADCOLA Soldering Instruments are of impeccable quality. You can depend on
ADCOLA day after day. That's quality. You can depend on
ADCOLA day after day. That's why they're so popular. You get consistent good service... relia-
bility... from our famous thermconsistent good service... relia-
bility ...from our famous thermally controlled ADCOLA Element
and the tough steel construction ally controlled ADCOLA Element
and the tough steel construction of $t$ his ideal production tool.

 nstruments are of impeccable


## for fast, easy, reliable soldering

Contains 5 cores of non-corrosive flux, instantly cleaning heavily oxidised surfaces. No extra flux required.

## SAVBIT ALLOY ALSO REDUCES COPPER BIT WEAR:

## Ecomically packed for

general electrical
and electronic
soldering. 75ft.
18 gauge on
plastic reel.
Recommended
retail price 75p
Size 12

A RANGE OF
SOLDERS IN HANDY DISPENSERS.

REF, ALLOY SWG
19A 60/40 88 18p* Size 5

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15 \quad 60 / 40 \quad 22 \text { 22p* }
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*Recommended Price

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 ESSENTIAL FORsoldering small components and thin wires. High tin content, low melting point, 60/40 alloy, 170 ft . 22 gauge on plastic reel. Recommended retail price 75p Size 10

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LTA 15 WATT AMPLIFIER
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Frequency Response $10-40,000$ cps-3dB. Bass Control- 17 dB

to -16 dB at 40 cps . Treble Control
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Recommended
£ 19

3-8-15 ohm Loudspeakers. Retail Price

PTA30 HI-FI
PUBLIC ADDRESS AMPLIFIER
A SUCCESSOR TO OUR
POPULAR CONCHORD
30 WATT UNIT
Retail Price
Size $9 t \times 3 \frac{3}{4} \times 5 \frac{1}{4}$ in. approx.


Output 30 watts.
Output Sockets for Loudspeaker or Recommended
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ohms. Three individually controlled
ohms. Three individually controlled inputs for mixing purposes. Housed in fully enclosed stove enamelle inputs for mixing purposes. Housed in fully enclosed stove enamelled
steel case. Controls Vol. (1) Vol. (2) Vol. (3) with mains switch. Treble 'lift'. steel case. Controls Vol. (1) Vol
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AN IDEAL UNIT FOR VOCAL AND INSTRUMENTAL GROUPS, SUITABLE FOR ANY KIND. OF 'MIKE" AND INSTRUMENT PICK-UP, ALSO FOR RADIO, TAPE OR GRAM.

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A compact system sold in matched pairs for a perfectly balanced stereo system. Each $\mathbf{t}$ enton con-
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A. new and exciting addition to the Wharfedale range, this three speaker system will satisfy the most ardent Hi i-Fi enthusiast. Shelf or floor standing-hand Frequency response $55=22000 \mathrm{~Hz}$ walnut. Dome pressure unit. Bass unit 8in. Mid. Dome pressure unit. Bass unit sin. Midlingedance $B$ ohms. Size 21 isin. $\times 9$ inin. $\times 9$ in.
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PRICE Teak Walnut


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High precision low-mass counterbalanced

Dick-up arm, heavy balanced turntable, simple to operate controls, viscous cueing
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A. Chassis only. B. Complete with Lasky's plinth and cover. $C$. Complete With Lasky's plinth, cover and AD76K cartridge. D. Comp. wired on BSR plint
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Digital Alarm Clock Radio.
The digital clock shows
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$5 \frac{1}{2}$ in. $\times 3 \frac{1}{2}$ in. fitted with
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LASKY'S PRICE $£ 9.95$


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Models HS-I and HS-2. Both these sets by TRIO offer really superb stereo reproduction in a lightweight, fuliy adjustable headset designed for oprimum comfort. Listening fatigue is Brief spec, both models. headphones. Brief spec. both models: Input imp.
$8 \Omega$ nominal (matching 4 to $8 \Omega$ ) input 0.5 W ; frequency response
$20-19 \mathrm{kHz}$
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Provides three separate switched output voltages
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| BF274 | 0.15 0.20 | Transistors |  |
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| $\mathrm{OCB}^{1}$ | 0.13 | AAY42 | 0.10 |
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This field effect transistor is the $2 N 3823$ in a plastic encapsulation, coded as 3823 E . It is also an excellent replacement for the 2 N 3819 .
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Complete stereo record and playback unit with line and microphone inputs. Fitted with tape counter, separate pause control, recording level metres for each channel,pop-up cassette ejection.Supplied complete with two pencil microphones. Wow \& flutter better than $0.3 \%$, frequency response $100-10,000 \mathrm{~Hz}$.Tape speed:17IPS,4•75 CMS.Rewind time: Better than 60 sec (C. 60 cassette).
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## SK 1

SOLDERING KIT
In rigid plastic "tool box" containing Model CW - 15 watts - 240 volts miniature iron fitted $\frac{3}{16}{ }^{\prime \prime}$ bit. Spare bits $\frac{5}{32}$ " and $\frac{3}{32}$ ". Reel of resin-cored solder, heat sink, cleaning pad, stand and booklet "How $£ 2.75$ to Solder".


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In polystyrene pack, containing 15 watt miniature soldering iron, 240 volts fitted with $\frac{3}{16}$ " bit, 2 spare bits $\frac{5}{32}{ }^{\prime \prime}$ and $\frac{3}{32}$ : ${ }^{2}$. Coil of resin-cored solder, heat sink, 1A fuse and booklet "How to Solder".


- 15 watts
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$£ 1.70$

Fitted with nickel plated $\frac{3}{32}{ }^{\prime \prime}$ bit and packed in handy transparent box.
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Battery-operated
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Complete with 15
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| ${ }_{185}$ | -28 | 25U4GT | . 57 | DY86 | $\cdot 26$ | EL500 | . 62 | PCL82 | - 38 | UABC80 | .88 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 185 | -22 | 30 Cl | -29 | DY87 | -26 | EM80 | -41 | PCL83 | - 59 | UAF42 | -51 |
| 174 | -16 | $30 \mathrm{Cl5}$ | . 68 | DY802 | - 40 | EM81 | - 41 | PCL 84 | -85 | UBC4I | $\cdot 59$ |
| 354 | -26 | 30C17 | . 80 | EABC80 | . 82 | EM84 | -32 | PCL85 | $\cdot 42$ | UBF80 | . 84 |
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| 5Y3GT | -28 | 30FL12 | $\cdot 70$ | EBC33 | . 40 | EZ40 | 48 | PENA4 | $\cdot 77$ | UCF80 | -84 |
| 5Z4G | - 85 | $30 \mathrm{FL14}$ | -70 | EBC41 | . 54 | EZ41 | . 48 | PEN36C | -70 | UCH44 | -68 |
| 6/30L2 | - 58 | 30 L 1 | . 30 | EBC90 | -22 | EZ80 | . 22 | PFL200 | $\cdot 58$ | UCH81 | -82 |
| 6AL5 | $\cdot 11$ | 30 L 15 | $\cdot 59$ | EBF80 | . 82 | EZ81 | -28 | PL36 | . 49 | UCL82 | - 84 |
| 6АM6 | $\cdot 18$ | $30 \mathrm{L17}$ | $\cdot 71$ | EBF89 | . 81 | GZ30 | -35 | PL81 | -48 | UCL83 | -55 |
| 6AQ5 | $\cdot 26$ | 30P4 | . 85 | ECC81 | . 17 | GZ32 | -42 | PL81A | $\cdot 49$ | UF41 | - 68 |
| 6at6 | . 22 | $30 \mathrm{Pl2}$ | $\cdot 77$ | ECC82 | . 20 | GZ84 | - 50 | PL82 | . 81 | UF89 | . 81 |
| 6AU6 | -21 | $30 \mathrm{P19}$ | -65 | EC083 | $\cdot 35$ | KT41 | $\cdot 77$ | PL83 | $\cdot 38$ | UL41 | -60 |
| 6BA6 | -21 | 30 PLI | -63 | ECC85 | . 26 | KT61 | -65 | PL84 | - 30 | UL44 | 1.00 |
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| 6BJ6 | . 41 | 30 PL 14 | -67 | ECF80 | . 30 | LN319 | -60 | PL504 | -64 | UM84 | . 28 |
| 6BW7 | -55 | 30 PL 15 | . 90 | ECF82 | . 26 | LN329 | . 72 | PM84 | -85 | UY41 | . 40 |
| 6CD6G 2 | . 07 | 35L6GT | -45 | ECH35 | . 30 | LN339 | -68 | PX25 | \$1.00 | UY85 | . 28 |
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| $6 \mathrm{~F}^{23}$ | - 70 | 35Z4GT | -25 | ECH81 | . 29 | P61 | 45 | PY33 | - 56 | Z77 | . 28 |
| 6F25 | - 60 | 807 | . 45 | ECH83 | . 40 | PABCS0 | $\cdot 34$ | PY81 | -27 | Transist | r: |
| 6K7G | . 12 | 6063 | .62 | ECE84 | . 86 | PC86 | .48 | PY82 | -25 | AC107 | $\cdot 17$ |
| 6K89 | $\cdot 17$ | AC/VP2 | . 77 | ECL80 | . 82 | PC88 | . 48 | PY83 | - 28 | AC127 | -18 |
| 6Q7G | -28 | B349 | -65 | ECL82 | . 81 | PC96 | -48 | PY88 | -33 | AD140 | . 87 |
| 6SN7GT | $\cdot 30$ | B729 | -68 | ECL86 | . 36 | PC97 | $\cdot 40$ | PY800 | - 84 | AF115 | -20 |
| 6V6G | -28 | CCH35 | -67 | EF39 | .88 | PC900 | . 38 | PY801 | - 34 | AF116 | -20 |
| 6V6GT | . 31 | CY31 | . 32 | EF41 | . 60 | PGC84 | . 80 | R19 | - 30 | AF117 | -20 |
| $6 \times 4$ | .28 | DAF'91 | -22 | EF80 | .23 | PCC85 | . 27 | R20 | -69 | AF118 | 48 |
| 6X5Gr'r | -28 | DAF96 | .36 | EF85 | . 28 | PGC88 | .42 | U25 | -65 | AF125 | $\cdot 17$ |
| $7 \mathrm{B7}$ | -38 | DF33 | . 38 | EF86 | . 81 | PCC89 | .46 | U26 | - 59 | AF127 | $\cdot 17$ |
| 10 Pl 3 | $\cdot 58$ | DF91 | -16 | EF89 | . 28 | PCCl 89 | . 48 | U47 | -65 | OC26 | -25 |
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| 12AT7 | . 17 | DH77 | . 22 | EF183 | . 28 | PCF80 | -29 | U50 | -39 | OC45 | - 12 |
| 12AU6 | -29 | DK32 | . 87 | EF184 | . 31 | PCF82 | . 81 | U52 | -31. | OC71 | - 12 |
| 12AU7 | -20 | DK91 | . 28 | EH90 | . 37 | PCF86 | . 45 | U78 | -24 | 0 C 72 | -12 |
| 12AX7 | . 23 | DK92 | . 42 | EL33 | . 55 | PCF800 | - 67 | U191 | -60 | OC75 | $\cdot 12$ |
| 19BG6G | . 87 | DK96 | . 38 | EL34 | . 45 | PCF801 | . 81 | U193 | 42 | OC81 | -12 |
| 20 F 2 | $\cdot 67$ | DL35 | . 40 | EL41 | . 54 | PCF802 | . 44 | U251 | -66 | OC81D | 12 |
| 20P3 | -80 | DL92 | -28 | EL84 | -23 | PCF805 | . 63 | U301 | 45 | 0 O 82 | . 12 |
| 20 P 4 | 482 | DL94 | . 87 | EL90 | . 26 | PCF806 | . 58 | U829 | $\cdot 66$ | OC82D | . 12 |
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Garrard SP25 Mk. III with CER. diamond
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£52.00
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SPEAKERS Duo Type II
Size $17^{\prime \prime} \times 10 \frac{3^{\prime \prime}}{4} \times 6 \frac{3^{\prime \prime}}{4}$. Drive unit $13^{\prime \prime} \times 8^{\prime \prime}$ with parasitic tweeter. Max. power 10 watts. 3 ohms. Simulated Teak cabinet. $£ 14$ pair $+£ 2$ p\&p.
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## SPECIFICATION R100/101

14 watts per channel into 3 to 4 ohms. Total distortion @) $10 \mathrm{~W} @ 1 \mathrm{kHz} 0.1 \%$. P.U.i (for ceramic cartridges) 150 mV into 3 Meg. P.U. 2 (for magnetic cartridges) 4 mV @ 1 kHz into 47 K . equalised within $\pm 1 \mathrm{~dB}$ R.I.A.A. Radio 150 mV into 220K. (Sensitivities given at full power). Tape out facilities; headphone socket, power out 250 mW per channel. Tone controls and fitter characteristics. Bass: +12dB to -17dB@ 60 Hz . Bass filter: 6 dB per octave cut. Treble control: treble +12 dB to $-12 \mathrm{~dB} @ 15 \mathrm{kHz}$. Treble filter: 12 dB per octave. Signal to noise rotio: (all controls at max) R101-P.U.I and radio-6SdB. P.U.2. -58dB. Rl00 same as RIO1 but P.U. 2 (for crystal cartridges) 450 mV into 3 Meg. Cross talk better than --35dB on all inputs. Overload characteristics better than 26 dB on all inputs. Size $13 \frac{3}{4}{ }^{\prime \prime} \times 9^{\prime \prime} \times 3 \frac{7^{\prime \prime}}{}$.

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enable the amplifier to be used at half power.
SPEAKERS! Size $20^{\prime \prime} \times 20^{\prime \prime} \times 10^{\prime \prime}$ incorporating $12^{\prime \prime}$ heavy duty SPEAKERS! Size $20^{\prime \prime} \times 20^{\prime \prime} \times 10^{\prime \prime}$ incorporating ${ }^{25}$ watt high flux, quality loudspeaker with cast frame. Cabinets atiractively finished in two tone colour scheme-Black and grey.

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| Q31 | 6 8il. mwitch trans. 2N708 NPN |  |
| -32 |  |  |
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| Q44 | 7 NPN trans. $4 \times$ BC108, $3 \times$ BCl | c.50 |
| Q45 | 3 BCI13 NPN T0.18 trans. | 0.50 |
| Q46 | 3 BCIIs NPN TO-ij tran | 0.50 |
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| Q48 | 4 BCY70 NPN trans. TO-18 |  |
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The price of 70 p applies only to catalogues purchased by customers in the U.K. and to BFPO addresses.

## Who makes what?

WELL, of course, that's easy! When you buy a radio set, hi-fi amplifier or piece of electrical equipment, the trade mark tells you that. Or if the mark is an unfamiliar one you can find the piece moulded on the cabinet or printed on the carton which says 'Made in Outer Mongolia' or whatever.

People are interested in where products are made for a variety of reasons. The possible standard of workmanship, value-for-money, reputation, reliability, the position of after-sales service, availability of spare parts-or even personal political motives-are all valid reasons why a purchaser would want to know who makes what, and where.

From November 30th, however, the Merchandise Marks Act comes to an end and from that date it will no longer be obligatory for imported goods to bear a country-oforigin' mark. The casual customer will therefore not be sure if he is buying a radio set made in Birmingham or Baluchistan, or an amplifier which is a Chinese-copy of an American design made in Hong Kong or Japan.

Of course, there will still be trade marks which proclaim that the goods are true-blue British or come from wellrespected foreign companies. Sadly, though, things are not always what they seem for many tape recorders, radio and TV receivers, bearing the names of old-established British manufacturers have for some time been made overseas and imported for the home market.

From November 30th, protection of the consumer has been thown at the retailer who, when asked, will be expected to disclose where a particular product is made. Unfortunately in may cases he will have no way of giving the required information. For instance, 'household-name' UK companies are none too keen to admit that some of their goods are being made in low-cost labour countries. Moreover, if a retailer does not know the country of origin, or (even inadvertently) gives the wrong answer, he will be at risk under the Trade Descriptions Act!

Apart from the factors mentioned earlier, which allow a customer to assess the advisability or otherwise of making a purchase, a particularly disturbing aspect of electrical goods (notably electric blankets and appliances) is that safety specifications may fall far short of those imposed on British-made goods and expose the user to potential hazards.

So, together with the British Radio Equipment Manufacturers' Association, the Radio and Television Retailers' Association and the Association of Manufacturers of Domestic Appliances, we beg the Secretary for Trade and Industry to think again. And to would-be purchasers after November 30th we can only say, most emphatically caveat emptor!
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Circuit Building Blocks
NOVEMBER ISSUE WILL BEPUBLISHED ON OCTOBER 8th

[^1]
## Mews... NEWS... NEWS...

Viscount value


The Viscount ill Audio System is the first of a range of audio equipment marketed in the UK by $\mathrm{R}+\mathrm{TV}$ Components (Acton) Ltd. under the Viscount brand name.

The complete system includes a choice of two amplifiers, a Garrard SP25 Mk III turntable and two Duo type II or type III speakers.

Amplifier RT100 is for ceramic cartridges and model RT101 which has an additional input stage accepts magnetic types. Both amplifiers incorporate f.e.t.'s giving excellent signal to noise ratios.

Prices are: RT100, £17; RT101, £22; Garrard SP25 with ceramic cartridge, $£ 21$ or with magnetic cartridge, £23. Duo type II speakers cost $£ 14$ the pair and Duo type III, f 32 .

Savings can be made by buying complete systems.

## Iripex cases

In last month's issue we stated the following about Impex instrument cases:
"At no additional cost, Impex cases can be supplied with all or some of these specifications: alternative colours of blue or bronze; no hood; no ventilating slots or back/front anodised front/rear panels".
What we should have mentioned is that these additions can onlybe supplied free of charge for quantities over 24.

## NHK's anniversary

June 1 saw the anniversary of the Radio Japan Service. Thirty-six years ago NHK commenced this service in commemoration of the 10th anniversary of radio broadcasting in Japan and the attainment of 2 million licensed subscribers.

From June 1934 to 1945, NHK broadcast its domestic programmes to Korea, Taiwan and Manchuria, then Japan's outlying territories, utilising telephone channels of the International Telephone Company. Shortwave broadcasts were received in various parts other than the areas directed, and reception reports
began to come to NHK from Japanese residing abroad and from large numbers of overseas people. In those days, Japanese living in various parts of the world totalled nearly two million. These people raised their voices requesting the establishment of an overseas broadcasting service. In Japan, too, commencement of an overseas broadcasting service was also strongly desired, since this country had been isolated after her withdrawal from the League of Nations and was in need of reinforcing its public relations abroad. At first the service was known as Radio Tokyo (Radio Japan was first used in 1952).

## Avo wall chart

Avo Limited (Thorn Group) announce a new wall chart featuring the latest Mark 4 version of the Avometer Model 8. The design and layout is such that it will prove of great interest not only to Avometer users but also to technical schools and general education establishments as a teaching aid.

Information is presented in a clear, logical manner and the theoretical and practical information is both concise and easy to understand. Magnetic theory is developed pictorially to outline the operation of the moving coil movement whilst a mathematical treatment is used to derive the various ranges of a multimeter.

Illustrations of the interior of the instrument include an exploded view of the movement to simplify the constructional detail and a circuit diagram provides an introduction to standard symbols and references.

The Avometer chart (size $20 \times 30 \mathrm{in}$., $508 \times 762 \mathrm{~mm}$.) is ideal for blackboard or wall presentation and is available from Spares Dept., Avo Limited, Avocet House, Dover, at 75 p per copy, including post and packing.



## METVS... METMS... METVS...

## Windermere VHF

The v.h.f. radio services from the Windermere relay station started on August 7. Radio 2 is on $88 \cdot 6 \mathrm{MHz}$, Radio 3 on $90 \cdot 8 \mathrm{MHz}$ and Radio 4 on $93 \cdot 0 \mathrm{MHz}$, all with horizontal polarization.

The v.h.f. radio transmissions from this new relay station, which is situated on Claife Heights, above the western shore of Lake Windermere, will serve Ambleside, Windermere town, Bowness-on-Windermere, and an area around the northern shores of the lake.

## Stereo from <br> Rowbridge

The necessary equipment has now been installed at Rowridge to enable the Radio 3 transmitter $(90 \cdot 7 \mathrm{MHz})$ to broadcast stereophony. This is being brought into use immediately, so that listeners can hear the Promenade Concerts in stereo-all except four of the concerts will be broadcast stereophonically in Radio 3. The stereo transmissions from Rowridge will initially be on an engineering test basis; every effort will be made to maintain them, but there may be occasions when it will be necessary to revert to monophonic transmissions. The date for the start of the full stereophonic service will be announced later.

## Miniature Electric Tools

A range of battery-operated miniature tools has been introduced by Expo (Drills) Ltd., 62 Neal Street, London, W.C.2.

Two basic models are available. The 'Reliant' is designed for lighter work such as model making, and has a full load current of 1.5 A . Its rated torque is $1 \cdot 38 \mathrm{oz}$. in. ( $100 \mathrm{gm} . \mathrm{cm}$.). For jobs requiring a more powerful tool and for professional applications the 'Titan Super' is rated at $3 \cdot 5 \mathrm{~A}$ on full load. It has a rated torque of 350 cmp operating at $4000-9000 \mathrm{rpm}$. An extremely large range of accessories is also available for both models. Various collets and accessories are supplied with the different models, which range in price from $£ 3$ to $£ 5 \cdot 50$.

## Eectures

THE Polytechnic of North London Department of Electronic and Communications Engineering is running a course on Audio and Acoustic Measurements on Thursdays 6.30-8.30 p.m. 1971/2. 1971: Oct., 28, Parameters and Standards of Measurement R. N. Baldock, B.Sc., F.B.K.S.T.S. Nov. 4, Measurement of Microphone Performance, H. D. Harwood, B.Sc., B.B.C. Research; 11, Electrical and Mechanical Measurements in Tape Recording, R. L. West, B.Sc., C.Eng., F.I.E.R.E.; 18, Measurements in Professional Recording, Angus McKensie; 25, Electrical and Mechanical Measurements on Disc, Pickups and Accelerometers, S. Kelly, C.Eng., F.I.E.R.E. Dec. 2, Measurement of Loudspeaker Performance I, R. L. West; 9, Measurement of Loudspeaker Performance II, R. C. Driscoll, M.Sc.; 16, Measurement of Studio Characteristics, R. C. Driscoll, M.Sc. 1972: Jan. 13, Measurement of Noise Transmission, J. Moir C.Eng., F.I.E.E.; 20, Assessment of Noise Nuisance, J. Moir, C.Eng., F.I.E.E.; 27, Visit to A.I.R.O. Laboratories, Hemel Hempstead, G. Berry, B.Sc., C.Eng., M.I.Mech.E. and Assocs.

## Grand Junk Sale

The Star Short Wave Club, G3ZWA, is holding a "Grand Junk Sale" in aid of Radio Amateur Invalid Bedfast Club funds. It is to be held at the New Inn Hotel, Bramley Town Street, Bramley, Leeds 13. The date is Wednesday September 15 th and time 7.30.

If any readers have any junk, the Star Club will collect within a radius of 25 miles from Leeds. Please contact Mr. T. Leeman, G8BUU, 115 Asket Drive, Seacroft, Leeds, LS14 1HX.

Chip "C's"


Featuring a very low insertion loss, a very high $Q$ and high capacitance-to-volume ratios, a new range of ceramic and porcelain chip capacitors from Guest International Limited is ideal for u.h.f. and microwave applications. Apart from the basic pellet and chip, there is a wide choice of lead variations: stripline, axial and radial wire, axial and radial ribbon.

ATC 100 series are porcelain chips with an insertion loss of $0 \cdot 03 \mathrm{~dB}$, a $Q$ of over 10,000 and a capacitance range of $0 \cdot 1$ to 1000 pF . In a case size of only $0 \cdot 110 \mathrm{in}$ square, the working voltage is up to 500 V d.c.

ATC 200 series are ceramic chips with an insertion loss of $0 \cdot 05 \mathrm{~dB}$. The capacitance range is 10 pF to $1 \cdot 2_{\mu} \mathrm{F}$ and chips are available in case sizes 0.050 , $0 \cdot 110$ and $0 \cdot 250$ in.square, depending on the value and the working voltage.



So many audio sources are now in common use that an article describing a specific design would meet the needs of only a few. This problem is overcome here by giving circuits for virtually all sources and care in design has made possible a large number of permutations so that a custom built audio mixer is possible even for those readers unable to design their own. The clever constructional method employed also allows for rapid modification to meet any mixing need.

ASTUDY of the four block diagrams in Fig. 1 will show how the various signal preamplifiers, the tone control unit and line amplifier etc. can be used to build audio signal mixing systems to cater for tape recording, public address, discotheque and even Hi-Fi applications. The combinations given are but a few of the permutations which can be extended for stereo use simply by duplication.

Each signal preamp has the same output level, 100 mV , and will match directly into either the tone control unit or the line amplifier, both of which are designed to accept a signal level of 100 mV . The tone control has unity gain so this provides 100 mV output to match into the line amplifier. Any group of signal preamplifiers can therefore be coupled directly to the line amplifier or via the tone control unit and then to the line amplifier. This allows any of the signal preamplifiers to be used with or without the tone control unit.
The line amplifier has a nominal output of IV r.m.s. at a fairly low impedance and can be used to drive a power amplifier module requiring between 500 mV to 1 V input, or it can be coupled to a tape recorder or radio input, or to a public address or music amplifier with inputs rated for 500 mV to $1 V$. Groups of the preamplifiers and line amplifier can be run from batteries or from an 18 V mains power supply.

A fairly simple arrangement is shown in Fig. la and is intended for use as a general purpose signal mixer as shown on the front cover. This would be suitable for tape recording, as it has two microphone inputs, one input for low impedance pick-up (with R.I.A.A. correction), and two line level ( 200 mV to over 1V) inputs. The tone control operates for the mic. and pickup inputs, whilst the line inputs go directly to the line amplifier. The system also includes a VU meter with variable control so that the meter can be set to read with the level meter on a tape recorder.

The combination shown in Fig. ib has the same input channels but no tone control unit. The arrangement in Fig. 1c might be suitable for a small discotheque with one mic. input, two pickup inputs both via a tone control unit and one line input, suitable for the output from a tape recorder. For music application, Fig. ld provides two mic. and two guitar inputs with separate tone control units for each guitar input. In each case the VU meter is optional but can be used as an overload indicator as will be shown later.

## PERFORMANGE SPECIFICATIONS

Seven preamplifiers will be described, each catering for specific signal sources as follows:-

1. LOW IMPEDANCE MICS. ( 25 to $600 \Omega$ ). Direct input, no transformer required. Input sensitivity $50 \mu \mathrm{~V}$. Output 100 mV . Frequency response $\pm 1 \mathrm{~dB}$. 30 to $30,000 \mathrm{~Hz}$. Signal-to-noise -50 dB . (Code: Mic. Lo-Z).
2. FOR MICROPHONES WITH BUILT-IN TRANS. FORMERS. Medium to high impedance. Input sensitivity 1 mV . Output 100 mV . Frequency response $\pm 1 \mathrm{~dB} 20$ to $30,000 \mathrm{~Hz}$. Signal-to-noise -60 dB ( Code: Mic. Hi-ZT).
3. FOR CRYSTAL MICS. High impendance. Input sensitivity 2 mV . Output 100 mV . Frequency response $\pm 1 \mathrm{~dB} \quad 30$ to $12,000 \mathrm{~Hz}$. Signal-to-noise -50 dB . (Code: Mic. Hi-ZX).
4. MAGNETIC PICKUP. Impedance $56 \mathrm{k} \Omega$. Input sensitivity 5 mV . Output 100 mV . Frequency response to R.I.A.A. as shown in Fig. 2. Signal-tonoise -60 dB . (Code: PU Mag.)



Fig. 2 : The RIAA response curve incorporated in the magnetic pickup module.
5. CERAMIC PICKUP. Input impedance $2 \mathrm{M} \Omega$. Input sensitivity 60 mV . Output 100 mV . Frequency response approximately 20 to $20,000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$. Equalisation automatic due to high input impedance. Signal-to-noise -50 dB . (Code: PU Cer.)
6. CRYSTAL PICKUP. Input impedance about $1 \mathrm{M} \Omega$. Input sensitivity 850 mV . Output 100 mV . Frequency response approximately 20 to $20,000 \mathrm{~Hz}$ $\pm 1 \mathrm{~dB}$. Equalisation automatic due to high input impedance. Signal-to-noise -60 dB . (Code: PU Xtal.)
7. GUITAR PREAMP. Input impedance medium, 10 to $100 \mathrm{k} \Omega$. Input sensitivity variable. 50 to 100 mV . Output 100 mV . Frequency response $\pm 1 \mathrm{~dB} 20$ to $30,000 \mathrm{~Hz}$. Signal-to-noise -50 to -60 dB . (Code: G.)

The tone control unit will take the output from any of the preamps and provides 100 mV output so that it can be coupled directly to the line amplifier. Its performance is as follows:- With tone controls neutral the flat response is 10 to $80,000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$. The response of the bass and treble controls is shown in Fig. 3 providing 15 dB lift and 12 dB cut at 40 Hz and 15 dB lift to 12 dB cut at $20,000 \mathrm{~Hz}$ The signal to noise performance is better than -90 dB . (Code: TC).

The line amplifier unit will accept the output from any of the preamps or the tone control unit and therefore has an input sensitivity of 100 mV . The output signal level available is $1 V$ r.m.s at an impedance of approximately $600 \Omega$. The frequency response is flat from 10 to nearly $100,000 \mathrm{~Hz}$ and the signal-to-noise performance -80 dB . The output can be fed simultaneously to one or more tape recorders and/or amplifiers via series output resistors as will be shown. The overall gain of the line amplifier is adjustable i.e., the input sensitivity can be increased. (Code: LA.)

The line inputs do not require separate preamplifiers as these are arranged in passive network coupling directly to the line amplifier. The input sensitivity will be 200 mV to over 1 V r.m.s. at approximately $10 \mathrm{k} \Omega$ impedance. (Code: $L$ input.)

## CONSTRUCTION

The construction of each of the preamplifier modules is such that the circuit board is mounted on a small screen this, in turn, being secured to the fixing bush of the gain control (VR1 in each circuit). The complete preamp can therefore be secured to a front panel by the lock nut of the gain control. The tone control unit is similar except that there are two potentiometers and both are mounted on the small screen and then secured to the panel. The line amplifier can be simply mounted on a piece of aluminium angle ( $3_{8} \times 3_{8}$ in) or on stand-off pillars as convenient. It need not be fixed to a panel but must


Fig. 3: Response curves of the tone control module.
be fairly well clear of mains power supply components. The input sockets for all inputs must be insulated from the front panel and earthed only at the common earthing point on the preamplifier board. If the mixing unit is to be mains powered, the components for the power supply must be kept as far away from the amplifier boards as possible and preferably with a screen around them.

The seven preamplifier boards are exactly the same and details for construction are given in Fig. 4. In each case the gain control (VR1) is fitted as shown

can be tested separately by operating them from an 18 V battery (two 9 V units) and by checking each one with an audio signal generator and audio voltmeter or by connecting an appropriate input signal source and feeding the output to an amplifier.

## LINE AMPLIFIER

The best order of construction is to build the line amplifier (code: LA) first so that the preamps can be tested with it as each is completed. The circuit is given in Fig. 6 and the board layout in Fig. 7. As


One of the sub chassis before the components are fitted; dimensions are given in Fig. 4.
with all the circuits except the tone control unit only one type of transistor is used-the BCl 09 silicon NPN-which is readily available and inexpensive. The line amplifier employs three of these beginning with $\operatorname{Tr} 1$ as an emitter follower input stage with $\operatorname{Tr} 2$ as the amplifier.

Negative feedback between Tr2 and Tr 1 via the pre-set PR1 ensures a linear response and some adjustment over total gain. The pre-set should be adjusted so that with a 100 mV input signal the output at C6 is exactly IV. If a signal generator and audio voltmeter are unavailable, set PR1 to midway position. The output transistor $\operatorname{Tr} 3$ operates as an emitter follower and provides a low impedance


An internal view of the prototype showing one of the modules. The input jack is on the left with the common wiring on the right.

Controls mounted as in Fig. 4 (c)
fig. 5 : Modifled sub-chassis for the tone control module.


Fig. 6: (above) and Fig. 7 (below): The circuit and layout of the line amplifier module.
output. Details for feeding more than one power amplifier and/or tape recorder from the line amplifier output will be given later.

## MICROPHONE PREAMP (Mic Lo-Z)

The circuit is given in Fig. 8 and the board layout


Fig. 8 (above) and Fig. 9 (befow): The circuit and layout of the low impedance microphone preamplifier module.

## components list




Fig. 10: Suggested circuit for a mains power supply to power the mixer.
in Fig. 9. The input transistor Tr1 operates as a grounded base amplifier and provides a very Iow impedance input. $\operatorname{Tr} 2$ is a signal amplifier with negative feedback introduced by R7 in series with the emitter by-pass capacitor C 4 . This resistor can be reduced a little in value to obtain more gain for very low output microphones, although it should not be less than $680 \Omega$ otherwise the noise level will rise excessively. As with all the preamplifiers the output is taken to the gain control VR1 and thence via the series resistor (R10) which becomes part of the passive mixing network common to all the preamps.

## MAINS POWER SUPPLY

The current consumption of the preamplifiers at


Internal view of the completed prototype.


The power supply should be properly screened as shown.
18 V varies between about 2 and 5 mA . The line amplifier takes 10 mA and the tone control unit 5 mA . Four preamplifiers, a line amplifier and say one tone control unit, would require about 25 mA which could be drawn from an 18 V battery with reasonable economy. Two PP9 batteries would be ideal. Larger mixing units, particularly if built for stereo use and requiring duplication of preamps etc. are best operated from a mains power supply. One that will provide well over 100 mA is given in Fig. 10 (see components list for transformer and rectifier details). The resistor R will be around $1 \mathrm{k} \Omega$ for low current drain and of lower value for higher current drain. With the mixing circuits running, adjust R to obtain 18 V across C 2 .

Part 2 next month, will deal with the rest of the preamplifier modules.

${ }_{4}$CHARLES MOLLOY THE NEW 1000 kW station on 1457 kHz located near Tirana in Albania is being heard in this country as a background to BBC Radio 4, (South and West). It relays Radio Peking and has been heard at 2100 hrs GMT with the identification 'Govarit Pekin' followed by programmes in Russian. An increasing number of superpower broadcasters are appearing on the medium waves. Urumchi in Western China ( 2000 kW ) is audible all evening on 1525 kHz ; listen between Prague 1520 kHz and Vatican Radio 1529 kHz for Russian speech plus jamming. Libya has two megawatters-El Beida on 1124 kHz and Tripoli on 1250 kHz . El Beida is the stronger of the two; it is heard with EAJ15 Barcelona on the same frequency throughout the evening. Tripoli suffers from interference from Dublin/Cork until 2245hrs when the latter close down, leaving the channel clear. Further east, Riyadh ( 1200 kW ) in Saudi Arabia is on 587 kHz and has been logged before midnight with Arabic programming.

Calcutta $1130 \mathrm{kHz}(1000 \mathrm{~kW})$ was heard frequently last winter before sign-off at 1700 hrs . It is now on extended schedule and has been several times during the summer at 2230 hrs GMT, as it changes from Chinese to Indian programming. There are three Voice of America stations in the Far East, each with 1000 kW , all of which have been heard in the U.K. Bangkok, Thailand 1580 kHz was logged at the end of September last year at 2200 hrs GMT while Poro 1140 kHz in the Phillipines and Okinawa 1178 kHz in the Ryuku Islands are both heard in winter before sign-off at 1700 hrs .
'Book of the World 1971' is the title of an unusual paperback now on sale in the bookshops. Sections deal with 181 different countries, each with a map plus details of the type of government, languages spoken, religions and currency. As well as providing the DXer with a valuable aid to station identification, this book gives background information on countries logged and adds interest to the hobby.

Broadcasts in English on the medium waves can be heard from a number of unexpected places during the late evening. Timisoara, Rumania on 755 kHz has a programme in English, French and German at 2230 hrs GMT, which should not be confused with the Voice of the West, in Portugal, also on 755, which starts its English programmes at 2245hrs. The Voice of America in Rhodes is 1259 kHz is heard regularly in English between 2100 hrs and midnight, with some interference from Wroclaw in Poland. Rhodes is one of the Dodecanese Islands situated near the coast of Turkey, which are usually counted as a separate country by DXers. The VOA is also heard in English from Thessalonika, Greece 791 kHz between 2100 hrs and 2130 hrs . Enugu, Nigeria 1320 kHz , although only 10 kW , is a weak but consistent signal at 2300 hrs GMT when it carries it's final news in English before closing down. ELBC Monrovia, Liberia on 629 kHz has English programming and is usually logged between 2330 hrs and sign-off at 0045hrs.

 which will go down as one of the ali-time 'greats', a 2-Octave Miniature Organ-complete with. voicing and tremulant. The electronics is dead simple (just look at the circuit sketch above!) and the cabinet woodwork can be bullt by anyone who can use з saw. The $£ 5$ cost includes everything-just as you see it.



## ALL IN THE NOVEMBER ISSUE ON SALE OCTOBER 8th

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## DIGITAL FREQUENGY GOUNTER/TIMER



## PART 2 (continued from the September issue)

This negative going step is passed by C5 to Tr 7 base which moves by -5 volts thus cutting off $\operatorname{Tr} 7$ and allowing the collector of $\operatorname{Tr} 7$ to rise to +5 V .

The current through VR2 and R14 discharges C5 so that the negative voltage on $\operatorname{Tr} 7$ base rises exponentially until Tr 7 conducts. The rate is determined by the setting of VR2, the DISPLAY control.

When $\operatorname{Tr} 7$ conducts, its collector voltage starts to fall, turning off $\operatorname{Tr} 6$ and thus increasing the base current to Tr7 and producing a fast negative going voltage at $\operatorname{Tr} 7$ collector. This voltage step is differen-

##     J.THORNTONLLAWRENCE GW3JGA

## NOTE

Several wiring connections were omitted from figure 6 in the September issue. Pin 10, on IC's 3, 4, 5, 6 and 7, should be shown connected to the OV rail. Pins 1 and 12 on IC's 9, 11, 13 and 15, should be shown joined.

On IC1; pin 4 Is the +5 V supply rail and pin 11 is OV supply rail. On IC2, pin 14 is +5 V supply rail and pin 7 OV supply rail. These latter connections were omitted from figure 6 for clarity.
tiated by C7, R16 and R17 and inverted by IC2d to give a positive pulse of $200 \mu \mathrm{~S}$ duration.

The positive going edge of this pulse resets the counter to zero and the negative going edge sets the JK flip-flop ICla and causes the cycle of operation to be repeated. The time for which the reading is displayed is controlled by the DISPLAY control and this is adjustable from about 0.5 second to 2 seconds.


| 0 | 0 | 0 | 0 | 0 |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 1 | 1 |
| 0 | 0 | 1 | 0 | 2 |
| 0 | 0 | 1 | 1 | 3 |
| 0 | 1 | 0 | 0 | 4 |
| 0 | 1 | 0 | 1 | 5 |
| 0 | 1 | 1 | 0 | 6 |
| 0 | 1 | 1 | 1 | 7 |
| 1 | 0 | 0 | 0 | 8 |
| 1 | 0 | 0 | 1 | 9 |


fig. 9:- Table showing the operation of the decade counter and decoder-driver.

## Gounter and Display

The counter and display has four decades, each decade operates in the same way. The output of one decade is linked to the input of the next higher decade to perform the 'carry' operation and a reset connection is common to all decades.

The input to the counter is a series of pulses. The counting has to be carried out in binary form and then decoded into decimal form for display.

This is done using a decade counter I.C. type SN7490N, which consists of 4 master-slave flip-flops forming a $\div 2$ counter and $a \div 5$ counter.

The input signal goes to the input of the $\div 2$ counter (pin 14) and the output of this (pin 12) is linked externally to the input of the $\div 5$ counter (pin 1).

The outputs of the four flip-flops are called $A, B, C, D$ and have a decimal value of $1,2,4,8$ respectively. The table showing the operation of the decade counter is included in Fig. 9.


Fig. 10: Exploded view of the function switch S2, viewed from the front.
The counter can be reset to zero at any time by the application of a positive pulse to pin 2 .
The binary coded decimal output from the counter, $\mathrm{A}, \mathrm{B}, \mathrm{C}, \mathrm{D}$ is connected directly to the inputs of the decoder/numeral tube driver I.C. type SN74141N. Here the inputs are decoded back into decimal form
for driving the numeral tube (Hivac Type XN13).
The numeral tube contains neon gas at low pressure and has an anode and 10 wire cathodes in the form of numerals, 0-9.

With the anode connected to HT + through a current limiting resistor, any cathode which is connected to 0 V will be surrounded by a glow showing up the shape of the number.

The outputs of the decoder-driver are collector circuits of high voltage transistors which, when made to conduct, effectively connect the appropriate cathode to 0 V . For example, suppose the input to the decade counter receives 5 pulses or cycles of input signal. The output will be D low, C high, B low, A high, representating a binary number of 0101. This number is decoded in the decoder/driver to figure 5 and the figure 5 output (pin 14) becomes switched to 0 V causing the figure 5 to be displayed in the numeral tube.

Each counter, decoder/driver and tube is built on a Veroboard panel and the inter connection of these panels is shown in the main diagram Fig. 6.

The decimal indication is provided by miniature wire ended neon indicator lamps mounted between two numeral tubes and these are switched by the function switch S2a.

## Function Switch

The function switch S2 is made up from Radiospares Miniature 'Maka Switch' parts and consists of a shafting unit and 4 break-before-make 1 pole 12 way wafers. The wafers are spaced by 2 spacers on each side. Only 11 positions are used and the stop on the indexing mechanism is adjusted accordingly.

The wafers are lettered a, b, c and d, starting at the rear and working towards the indexing mechanism. Wafer ' $a$ ' controls the decimal point neon lamps, wafer ' $b$ ' the clock period input to the gate, wafer ' $c$ ' the signal input to the gate and wafer ' $d$ ' the output from the clock divider.

In the frequency positions 1 to 5 , the output from the input stage is fed to the 'Signal Input' of the gate and the appropriate clock signal is fed to the 'Clock Period Input' of the gate. In the time positions, 7 to 11 of the switch, the output from the input stage is fed to the 'Clock Period Input' of the gate and the appropriate clock signal is fed to the 'Signal Input' of the gate.

A test position is provided in position 6 where the 1 kHz clock output is counted for a period of 1 second to give a reading of 1,000 . The test position only serves to check that the gate, counter and display are functioning, it does not provide a check of the crystal oscillator frequency.

The wiring and connections to the switch are shown in Fig. 10. The method of assembly is as follows:

1. On the shafting unit, check that the lengths of studding are screwed fully into the indexing mechanism plate and remove the end nuts and washers.
2. Rotate the switch spindle to the fully anti-clockwise position (as viewed from the front). The flats on the wafer end of the spindle should be in line with the studding. This is position " 1 ".
3. The switch should now be held so that the lengths
of studding are at each side and the control knob end of the spindle is towards you, as shown in Fig. 10. 4. Take a 1 pole 12 -way wafer and hold this with the side having the majority of tags, away from you and with the two tags that are facing you positioned in the top right hand segment. Rotate the rotor section until the common or rotor contact is making contact with tag 1 . The slot in the rotor should now be horizontal.
4. Assemble the wafer on to the shafting assembly, keeping the side with the majority of tags facing away from the indexing mechanism. This wafer is now wafer "d" in Fig. 10.
5. Fit 2 spacers on each piece of studding.
6. Fit another wafer as described in operation 4 and 5 (wafer "c").
7. Repeat operations 6 and 7 until four wafers have been fitted (wafers "b" and "a").
8. Fit washers and fixing nuts to the studding.
9. The rotation of the switch should now be checked to see that all parts operate smoothly.
10. The rotatable stop plate, which is located on the face of the indexing mechanism, should now be set to give 11 positions, working clockwise and commencing from the maximum anti-clockwise position. This is the position at which all the wafers have been set. The stop plate may be held in place temporarily by the mounting nut, ready for assembly to the front panel.
11. The wiring of the switch may now be carried out as shown in Fig, 10. 22 s.w.g. tinned copper wire is used for wiring from tag to tag, sleeving being necessary only when crossing between wafers.
12. The correctness of the wiring can be checked by using an ohm or continuity meter and measuring between the appropriate tags whilst the switch is rotated, e.g. $S 2 b / C$ is connected to $S 2 d / C$ in positions $1,2,3,4,5$ and 6.

## Power Supply

The power supply consists of a Radiospares Midget Mains transformer which with its associated rectifiers smoothing and regulators provides

$$
\begin{array}{ll}
+300 \mathrm{~V} \text { at } 10 \mathrm{~mA} & \text { Unregulate } \\
+5 \mathrm{~V} \text { at } 350 \mathrm{~mA} & \text { Regulated } \\
-5 \mathrm{~V} \text { at } 5 \mathrm{~mA} & \text { Regulated }
\end{array}
$$

The output from the 250 volt secondary winding is rectified by D 7 and smoothed by Cl 2 to give about +300 V output. The negative return is through R24 and D9 to 0 V . The negative supply derived in this way is smoothed by C13 and regulated by the zener diode D9 to give -5 V . The I.C. logic requires about 350 mA at +5 V and this is provided by the 6.3 volt flament winding on the transformer.

The $6 \cdot 3$ volt output is full wave bridge rectified to give about $+7 \cdot 5$ volt across the reservoir capacitor C14. This feeds the zener diode D6 and the power emitter follower Trll to give the required 5 V output.

## Veroboard Panels

The layout drawings of the $0 \cdot 1 \times 0 \cdot 1$ in matrix Veroboard show the copper strip side of the panel the components on the reverse side being "seen through" the copper strips.

Each copper strip is given a letter code and each


Fig. 11. Veroboard panel A. Input stage, gate and control oscillator. Viewed from copper side. Note correct orientation of integrated circuits when fitting. The integrated circuits and their holders are indented at one end to assist orientation.
horizontal row of holes, a number.
All solid horizontal lines represent wire links which connect one strip to another.

The orientation of integrated circuits and transistors should be noted carefully, some transistors may have unusual lead positions and these may have to be crossed over for the connections to be correct. Assembly of the components, etc., must be carried out in a systematic manner or errors will result. A suggested method is as follows:

1. Cut the Veroboard to the correct dimensions using a fine toothed saw and working from the copper side. Apart from checking the physical dimensions, also check that the correct number of holes appear in each direction.
2. Fix strips of masking tape on all four edges of the Veroboard panels. Also fix masking tape to the same edges on the reverse (copper) side of the board.


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8.75.
Carr. \&1. 5.75. Carr. \&1. MASTS $5^{\prime} \mathbf{2}^{\prime \prime}$ Extension sections to
fit bottom of above fit bottom of above mast to
increase height. increase height. sl-25 each
(any number sapplied).

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Fig: 12. Veroboard panel B. Crystal clock and divider, Viewed from copper side
3. By writing on the masking tape, letter all the vertical strips and number all the horizontal rows of holes on both sides of the board.
4. The breaks in the copper may be made using the appropriate Vero tool, part No. 2022 or by using a $\frac{5}{16}$ in. drill held in a pin vice. Each copper break on the layout drawing should be identified by its co-


Fig. 13: Counter and display pane/s C. D. E. F. Panel F. does not require a neon "decimal".
ordinates, e.g. D38 and the copper strip removed by carefully rotating the tool or drill by hand using the identified hole as a guide.

Experience has shown that just counting the position of a hole is not reliable and errors are very likely to result.
5. Wire links are formed from pre-stretched $22 \mathrm{~s} . \mathrm{w} . g$. tinned copper wire. A length of about 3 ft of $22 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. TC wire is stretched slightly by using a pair of pliers at each end, until it becomes stiff. All wire links are fitted on the components side.

A length, in excess of the length of the link required, is cut off and two right angled bends are made in the appropriate places to form the required "U" shaped link. A pair of pointed nosed pliers is useful for this and the span of the link can be checked by holding it against the panel in the appropriate place. When the link is inserted, the end wires may be bent outwards slightly to hold the link in place while it is soldered.
6. Components should be fixed starting at one end of the panel and the position and value checked before soldering.
7. The soldering should be carried out using a soldering iron with a small chisel shaped bit, no wider than $\frac{1}{16} \mathrm{in}$. The cored solder used should be no larger than 20 s.w.g.

Check each soldered joint using a magnifying glass as it takes only a very small piece of swarf
or solder to bridge the $0 \cdot 025$ in gap between the copper strips.
8. Connections to the Veroboard panel are made to pins inserted in the appropriate places and soldered to the copper strip. These pins are Vero type TP 11032 or similar.

An insertion tool for this pin is Vero part No. IT 2151.
9. Tick off each break, link, component and pin on the drawings as they are fitted, this will identify any items which otherwise might be missed.

## Semiconductors

The 3702 and 3704 transistors are specified in the components list as being the ' 2 N ' types, these are the most popular version and are readily available.

The 3702 and 3704 transistors are also available


Fig. 14 (/eft) Cabinet and chassis construction.
in an ' $S X$ ' version. The lead connections for these types are all different from each other and are shown in the diagram above.

It should be noted that on the Veroboard panels, the connections run e.b.c. and when the ' $2 \mathbf{N}$ ' version is used, the collector and base leads will have to be crossed over to line up with the appropriate holes in the Veroboard panel.

## Sockets for Integrated Circuits

Sockets are used for all the integrated circuits and although this increases the overall cost and is not essential for the operation of the instrument, it does provide the facility for disconnecting or changing an I.C. should a fault condition exist, also it does allow various parts of the instrument to be tested without incurring the expense of buying all the I.C's at once, 16-way sockets are required for I.C.8, 10, 12 and 14 and 14-way sockets for all the others.

The instrument is built using a simple aluminium chassis, panel and cover type of construction, as shown in Fig. 14. The finished chassis size is $8 \times 7 \times{ }^{3} \sin$ deep and has a front and rear panel, each size $8 \times 3$ in, fastened to the front and rear edges with 6BA screws and nuts.
(To be continued)

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| IEsenticer |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Code | Power | Tolerance | Range | Va |
| 0 | 1/20W | 5\% | $82 \Omega-220 \mathrm{~K} \Omega$ | E |
| C | 1/8W | 5\% | $4.7 \Omega-470 \mathrm{~K} \Omega$ | E |
| C | 1/4W | 10\% | $4.7 \Omega-10 \mathrm{M} \Omega$ | E |
| C | 1/2W | 5\% | $4.7 \Omega-10 \mathrm{M} \Omega$ | E |
| C | 1w | 10\% | $47 \Omega-10 \mathrm{M} \Omega$ | E |
| MO | 1/2W | 2\% | $10 \Omega-1 \mathrm{M} \Omega$ | E |
| WW | 1W | $10 \% \pm 1 / 20 \Omega$ | $0.22 \Omega-3.9 \Omega$ | E |
| WW | 3W | $5 \%$ | $12 \Omega-10 \mathrm{~K} \Omega$ | E |
| WW | 7W | 5\% | $12 \Omega-10 \mathrm{~K} \Omega$ | E |
| Coden: $\quad 0=$ carbon flm high stability low noise |  | flm high stabil | ty low noise |  |
| $\begin{aligned} \mathrm{MO} & =\text { netal oxide Electrosil Ths ultra low noise } \\ \mathbf{W W} & =\text { wire wound Plessey. }\end{aligned}$ |  |  |  |  |

## Values:

Et2 denotes series: $10,12,15,18,22,27,33,39,47$, 66 , 68, 82 and their decades.
E24 denotes series: as $\mathrm{ER2}$ plus 11, 13, 16, 20, 24, 30,36 , $43,51,62,75,91$ and their decades.

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$0.047,0.056,0.068,0.082,0-1.0 .12,0.15,0.18,0.22,6 \mathrm{p}$ $0 \cdot 27,7 \mathrm{p} ; 0.33,9 \mathrm{p} ; 0.39,9 \mathrm{p} ; 0.47,10 \mathrm{p} ; 0.56,18 \mathrm{p} ; 0 \cdot 68,15 \mathrm{p}$

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| Audio |  |  | 5 -pole 180 deg. | 15p | 12 p |
| Audio | . | . | 5 -pole 240 deg. | 15p | 12p |
| Audio | . |  | 6-pole | 15p | 13p |

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THE technical press offer so many transistor projects that many of us have bought and acquired a fair stock of transistors gathered from bargain packs or labouriously removed them from surplus computer boards. We then go through them only to find that many are un-marked, indecipherable, or have some queer markings not contained in any lists. The manufacturers are reluctant to supply, data because they have been made to "customer requirements". A few hours spent in sorting these into NPN or PNP and grading them into four categories will save a lot of time, patience and money on the next project development. The methods of grading are only approximate and there are only four grades: -

1. RF types.
2. Low gain types.
3. High gain types.
4. All others which do not pass the tests described.

## PRELIMINARY TESTING

First do a thorough visual examination and sort the transistors into known and unknown. The knowns, however, should be put through the tests since they may be blown or below spec. If your collection contains transistors removed from computer boards, the leads will be short. Fortunately when these boards are assembled, small plastic

Fig. 1: Some transistor lead connections. These are only the most common arrangements and variations do exist.

spacers are fitted, in most cases to comply with manufacturer's recommendations as to bending and heat dissipation. The makers also usually bend over about $I_{8}$ in on the solder side with the result that the lead lengths are in a safe condition. Any that are too short should be thrown away. Figure 1 shows the bases of seven categories into which these recovered units fall. Some may have a coloured dot, usually red or white, indicating the collector lead, also some are internally connected collector-to-can. Do not be


Fig. 2 : Resistance readings found on a good transistor-see text for details.
fooled however by coloured sleeving on the leads; not all manufacturers use the same code. Of course a really good transistor checker can make things very easy but this will cost a good deal of money.


## IDENTIFYING THE LEADS

The first checks, sorting the PNP and NPN types and the leads are done with a multimeter on its low ohms range. A word of warning first about that multimeter. Most meters on the ohms range give positive volts at the NEGATIVE LEAD and negative at the POSITIVE lead, so check first. Figure 2 illustrates the method of finding out whether the transistor is PNP or NPN. Never be scared of reversing the leads from the meter; with about 1.5 V , which is usually
available from the meter on low ohms ranges, you won't do any harm. Should you get a low resistance reading both ways between base and either of the other connections you can be fairly certain that there is an internal short. If you get a low reading between the base and one lead and a high at the other it is either bad or you are not on the base. It is probable that you will meet with marginal readings at times, so make tests as follows with reference to Fig. 3.

'scope, all the better. Now adjust VRl until a tone (or trace) is observed, then start to substitute the unknowns for Trl. If a tone is present the gain will be more than 30 , if no oscillation occurs then it is less than 30. At this point those transistors which have connections which are difficult to identify can be checked and identified by insertion, keeping the assumed base connection and reversing the other two unknown leads in the socket. If no oscillation occurs, keep it for the two further tests. Do not be afraid to reverse leads even at this higher voltage; if it blows then it was probably useless anyway.

## AMPLIFICATION TEST

Now replace the original test transistor i.e. the 0C44/45 (or one of the good ones known to oscillate) and then one by one test all those which failed the high gain test by substitution in the Tr2 position. The note heard in the headphones will tell you if it will amplify at all. If it is possible to use an oscilloscope, a little more information will be gained on this last test which is for low gain. By means of the trace
Fig. 3: Tests for identifying the base connecilons.
that for NPN devices the battery polarity must be reversed.

Draw a base diagram of the base of the transistor and mark them ' $A$ ', ' $B$ ', ' $C$ '; you are now faced with three permutations, of pairs ' AB ', ' AC ' ' BC '. Now take any pair and connect to the ohmmeter on low ohms range, then connect the other lead to each lead in turn: if the resistance falls the unpaired lead is the base lead and the lead to which the base was shorted is the collector. Check to see if this lead is connected to the metal can, this could confirm that it is the collector. If you do not get these results then check each of the three pairs using the same procedure. If the base cannot be identified for certain and there is high resistance both ways it is most probable that the third lead is base. To check, assume this is base and check with the meter whether it is a PNP or NPN in Fig. 1. You should then have established the base connection; the emitter and collector connections will be identified by reversing the leads in the grading test.

## TESTING FOR GAIN

To test for gain two simple circuits are required as shown in Fig. 4. Good quality transistor sockets are essential since we are most likely dealing with short leaded transistors and soldering in and out will become tedious. The assembly of the the circuits is an individual choice, though the author used ' S -Dec' and found it ideal. When the test rig is assembled, put in two good known transistors, say an OC44 or OC45 for Trl and an OC70 or OC72 for Tr2, connect up the battery and headphones. If you can use a

Fig. 4: The circuit used for the gain and amplification tests. Note
 shown on the 'scope, you can classify some transistors in the large signal class because the waveform will be distorted or clipped. The assumption in the tests up to now has been that the surplus transistors are in the small signal group.

## R.F. TESTS

All the transistors available should be put through the r.f. test, irrespective of whether they have passed the previous tests. The r.f. test circuit is shown in Fig. 5. Here the reader will have to arrange for the crystal, Trl, L1 and C1, according to what he has available, to check frequency. As a guide, with a 20 MHz overtone crystal, an $0 \mathrm{Cl} 70 \mathrm{~L}, \mathrm{~L} 1$ being eight turns of 18 s.w.g. on a ${ }^{1}{ }_{2}$ in former, tuned with a 50 pF capacitor. the circuit will cover the range up to


Fig. 5: The r.f. test circuit. For NPN devices the battery polarity must be reversed.

30 MHz . Here a receiver or G.D.O. is necessary to detect the oscillation. Once oscillation is established it is merely a matter of substituting as before.

With your collection graded, as outlined, it is simply a matter of substitution of various types within that grade for optimum results in a particular project. The methods described are by no means comprehensive but at least you will have got a little order out of chaos and the building of many projects will be made much easier.

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6 in . 9 in . and 12 in . miniature tubes kl .
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$\begin{aligned} & \text { Postage on Kits A and B 23p for one or two } \\ & \text { kits then } 23 \text { for each two } \& \text { its ordered. }\end{aligned}$
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published.

# Gexirkil purpose 



IN the "good old days" of valves, power supplies were simple. Some 90 per cent of equipment operated with an h.t. supply of about 300 V and the heaters were nearly all $6 \cdot 3 \mathrm{~V}$. A single power supply would operate the vast majority of equipment that one was experimenting with. Today the semiconductor rules the day and power supplies are very much less standardised. Checking through a recent copy of Practical Wireless showed that there were six circuits given-and five different supply voltages were specified, ranging from 6 V to 30 V . Batteries can cope with a number of these, especially the most common 9 V supply, but even the most half-hearted experimenter needs quite a range of batteries to pursue his hobby-even for quick lashups.

One can of course build a sophisticated, stabilised power supply with variable output covering quite a large range but such circuits are expensive, are by no means simple in their construction and most of the time the sophistication is not required.

By incorporating a simple two-pole, six-way rotary switch into the circuit of a standard power supply, it is possible to build a power unit with six different outputs-all with low internal impedanceand which will simulate exactly the supply that you finally build into the equipment.

A few refinements can be incorporated into such a power supply such as an indicator neon and fuse protected output which are not always necessary on the final supply which is fitted permanently but which make life simpler (and cheaper!) while experimenting.

## the circuit

The wide proliferation of supply voltages needed for semiconductor designs has brought about the introduction of mains transformers with a wide range of secondary tappings and it is simply by selecting the appropriate ones of these that various output voltages can be obtained.

The transformer specified in the components list is a good one to use but other types with multiple secondaries can be adopted.

This one is fitted with $12,15,20,24$ and 30 V tappings (these of course refer to a.c. r.m.s. volts)


Fig. 1: The circuit of the general purpose power supply.


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experience that the secondary taps are by no means always accurate and quite large differences have been measured from the theoretical voltage but as exact voltages are rarely needed, variations are allowable.

The output voltage from the prototype, which can be seen marked on the chassis, are those measured, not those calculated, and these discrepances can only be due to winding inaccuracies as it is wired up exactly as in Fig. 1.

Fig. 2: The component layout and wirling details.


The current rating required from a general purpose power supply will depend on the uses to which it is put. Only power amplifiers at peaks take more than about 500 mA and that was the rating chosen; it is sufficient for nearly all experimental work.
More than one company makes transformers with the voltage taps used here, but the most widely available is possibly the Douglas MT112AT. The same transformer with a 2A rating is coded MT3AT and although it will just fit into the chassis shown, it is a tight fit and a larger one is recommended.
The mains supply is connected to the correct primary (either 210 V or 250 V ) of the transformer via a double-pole switch. A panel neon, connected


Fig. 3: The switch wiring used on the prototype.
across the primary is very useful. Today most neons are supplied with a built-in resistor but if one is not fitted, a $220 \mathrm{k} \Omega$ should be wired in series as shown in the circuit.
The earth line is shown connected to the negative line but may equally well go the positive if used in conjunction with PNP type equipment which is earthed by some other means. Whatever happens the aluminium chassis should be earthed for safety reasons.
The various connections to the voltage selector switch are taken from the appropriate tappings and the switch wipers are connected directly to the bridge rectifier a.c. contacts. Plastic encapsulated silicon bridge rectifiers are widely available at reasonable prices and come under a variety of codings. The ratings must be 50 V peak, 1A. The Mullard BY164 is suitable here.

Adequate smoothing for all purposes is provided by the $2000 \mu \mathrm{~F}$ capacitor; the voltage rating of this should be well above the 45 V or so applied and a 64 V type was used in the prototype.

To provide a measure of protection, a 500 mA fuse should be connected in the supply as shown.

## construction

The prototype was built in a
small aluminium chassis size $5 \times 4 \times \mathbf{2 1}_{2}$ inches, fitted with a drop-in lid. These chassis are available from H. L. Smith Ltd., the address is given in the components list.

The layout of the components is shown in Fig. 2 and can be seen in the photograph.
The bridge rectifier can be mounted on a small tag strip on one long side. the transformer and the


Fig. 4: The front panel drilling detalls.
capacitor on the base, with the remaining components on the front panel.

The switch wiring needs quite a lot of planning unless that used in the prototype is copied exactiy, this is shown in Fig. 3. If other voltages, selected from the table, are required, the wiring should be planned carefully first of all. It is far easier to do the wiring on the switch leaving a few inches of wire which will finally be wired to the transformer.
When all wiring is completed a thorough check should be made to see that there is no possibility of


Internal view of the prototype showing component siting. Compare this with Fig. 2.
mains being applied to the case and that no bare wires can touch each other.
Connect a voltmeter to the terminal sockets and switch on, ensuring that a voltage reading is obtained in each switch position. Wrong wiring on a switch of this sort can lead to a direct short of the secondary and-unless noted very quickly-unpleasant things can happen to both the transformer and the switch.

There may be a very small amount of sparking while switching but this does no harm. The prototype has been in use continuously for several months with no ill effects.

There is just one point that must be made. The charge stored in a smoothing capacitor of this size lasts for a long while and if the power supply has been on a high range the capacitor should be shorted out (with the mains switch off) before being applied to a circuit where a high voltage could cause damage. This should be done through a resistor, not directly, as the fuse may blow otherwise.
If, due to the nature of the experimenting, a considerable amount of switching is envisaged, a resistor can be wired permanently between the output terminals so that a small current is drawn at all times and the capacitor will then settle down very quickly to the applied voltage. A $4 \cdot 7 \mathrm{k} \Omega$ resistor with a ${ }_{2} \mathrm{~W}$ rating will serve this purpose.

## CROSSWORD No. 2 WINNERS and SOLUTION

The Practical Wireless Crossword Competition No. 2, published in the June 1971 issue, has now closed and the prizes of $£ 10$ vouchers, exchangeable with any advertiser in Practical Wireless have been awarded to:

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# SIIIPLL UORRSHOP TOOLS s.veaman 

SINCE YOU ARE reading Practical Wireless the chances are that you have already tried your hand at constructing one of our projects. If you are already well fitted out with tools or have access to workshop facilities the bending up of a piece of aluminium to make a chassis or the punching and drilling of various holes would not have presented you with much of a problem. But if you are a newcomer to home electronics you have probably employed 'knife and fork' techniques, utilising kitchen knives and similar tools!

This really is doing things the hard way and the effort involved can put a considerable damper on anyone's enthusiasm for home construction, apart from any ensuing domestic problems when the lady of the kitchen discovers that her best kitchen knife now looks like a hacksaw blade!

If you are really taken with the idea of constructing electronic gear then be prepared to start off properly by buying a few good essential tools designed for the job in hand. Forget Dad's woodworking screwdriver with its $\frac{1}{2}^{\prime \prime}$ wide blade filed so thin that it seems to fit any screw head from 8BA upwards! Very convenient, of course, but what a mess it makes of screw heads! Dad's pliers, too, are very handy-they 'fit' any size of nut you like to mention but by the time they have slipped a couple of times that nut looks pretty sorry for itself.

When you find out what fun these electronic projects can be you will begin to take pride in the finished job and Mum's knives and Dad's tools just won't be good enough. So, start off on the right foot by getting the proper tools and remember that, like most 'bargains', cheap tools can turn out to be expensive in the long run. Spend a little more on tools from manufacturers of repute and scorn imported tools made of inferior materials.

The following review lists small workshop tools more or less in the order of necessity bearing in mind the requirements of the home constructor. The soldering iron takes precedence since even the simplest bought kit of parts needs the odd soldered joint or two.

In a review of this kind only general reference can be made to the great variety and range of tools that are available today. Specific recommendations are impossible without some knowledge of the amount of money it is intended to spend and the precise nature of the work to be undertaken in the workshop.

Workshops tend to 'just grow' from a few simple tools. As the limitations of these tools become apparent so more tools are bought to fill the gaps, thus it can be seen that a large initial outlay is not necessary.

Spreading the cost over a period allows emphasis to be made here on the requirement to buy quality, a policy which, if followed, will not be regretted.


SOLDERING IRONS. There is a wide variety of electric soldering irons available today, the main problem being the selection of one of a suitable wattage for the home constructor of radio gear. For the average Practical Wireless project a 25 watt iron should prove satisfactory but if a lot of work is likely to be done on printed circuit boards a 15 watt iron would be more appropriate. These circuit boards often contain many small components fitted quite close together and thus vulnerable to the heat of a soldering iron. Excessive heat can also damage the copper strips of the board itself.
With these irons goes a wide range of copper bits of differing diameters, some straight and some angled. Initially a straight bit $\frac{1}{8}$ " or $\frac{3^{\prime \prime}}{16}$ in diameter can be chosen for general purpose work with an angled bit for the awkward spots. At a little extra cost special bits are available which contain additives to the copper designed to prolong the life of the bit.

For home use a soldering iron operating directly from the 230 volt mains supply is usual but, for anyone who might feel a little happier with a lower working voltage, irons are a vailable to work from as low as 6 volts via a mains is olating transformer.

When wiring an electric iron to its mains plug ensure that the instructions supplied with the iron are strictly adhered to so that the iron is properly earthed. This is a very important point for the builder of electronic equipment when wiring in transistors and other solid-state devices. Irreparable damage can be caused to these devices by stray earth currents if the soldering iron is not properly earthed.

Advancing rapidly in popularity are the soldering guns or 'heat guns', as they are sometimes known. The copper bit is, in fact, part of a low voltage secondary winding of the built-in transformer. Exaggerated claims are sometimes made concerning the short heat-up time for these guns,

electronic soldering. Various tips are obtainable in copper or long life Permatip. (Light Soldering Developments Lid.)

In the Adamin series of miniature irons the copper bit fits over the element. The smallest, Model 6, has a rating of only 6 watts at 6 volts. (L.S.D. Lid.)


This soidering iron bench stand guards against accidental burns or even fires. The iron maintains its normal operating temperature while in the stand. A bit cleaning pad is fixed to the front of the stand. (L.S.D. Ltd.)
typically 10 seconds, but the better known makes can be relied upon in this respect.

The trigger style on-off switch on these guns often has two positions, for fast heat-up and for running heat, thus giving a dual-heat facility. Sometimes other types of bit are available which can be used to cut sheet plastics or to weld plastic sheet and so on.

Solder today, as used on electronic work, invariably has several cores of flux to ensure better distribution of the flux around the joint and consequently a better joint. The solder contains tin and lead in varying proportions which affect the temperature at which the solder will melt. The usual ratio of tin to lead is $60 / 40$ for radio work and the diameter of the solder can be between 18 and 22 s.w.g. depending upon the fineness of the work being done.

While short lengths of solder can be bought for a few pence the $\frac{1}{2}$ or 1 lb . reel will be found to be more economical for the home constructor. If however he wishes to experiment with various solders he will find many easy-to-use dispensers and handy-packs available with different types of solder such as $40 / 60$ or $60 / 40$ tin-lead alloys as well as solders containing a small amount of copper which is calculated to reduce the amount of wear on the copper soldering bit.

Although soldering irons are usually fitted with a hook of some kind so that they can be hooked onto a convenient point it is a good policy to provide some kind of metal stand for use on top of the bench. Such a stand can be easily made out of an odd piece of aluminium or tinplate.

Soldering iron manufacturers naturally produce their own stands or 'protective shields' for their irons and these are designed to ensure that the iron does not overheat when the iron is not being used but is still switched on. For convenience a cleaning pad for the bit may be incorporated and possibly a holder for a reel of solder.

A 'must' in soldering is a sharp pick or long needle for clearing out holes in soldering tags and removing surplus solder. Commercial versions are usually made of aluminium, to prevent the solder sticking to the needle, and with a plastic handle or covering. One end may be pointed and the other end split to enable wires to be unravelled from tags etc., while the solder is molten.

Another de-soldering aid available is in the form of copper wire braid. When the end of the braid is touched on a hot soldered joint surplus solder travels up the braid by capillary action. The end of the braid is snipped off ready for the next joint.


PLIERS AND CUTTERS. Again a very large field of choice and again the rejoinder to pay a bit extra to get quality. Two pairs of pliers will suffice for most constructional work, a long-nosed pair with tapered jaws coming to a reasonable fine point and a heavier pair with stocky jaws, maybe $\frac{1}{2}^{\prime \prime}$ wide, and probably incorporating a pair of sidecutters.

The long-nosed pliers will always be in the hand of the electronics constructor, for twisting wires round tags and forming the wire ends of components before soldering them into position. Don't forget that a wiring joint should always be mechanically sound before it is finally soldered, although the true experimenter will not go along with this principle because he will want to use the same components time and again!

Variations on the pliers theme include those with angled
jaws of various lengths, round jaws and flat narrow jaws rather than pointed ones. When buying pliers see that the opening and closing action is not too loose or the jaw tips will not close together properly and it will be difficult to get hold of fine wires when wiring up.

Wirecutters will be used as often as the pliers so, again, only the best must be chosen. They should be about the same size as the long-nosed pliers with the cutting edges at an angle to the body. Other types have the cutting edges across the top or parallel with the body.

When buying sidecutters hold them up to the light and see that the cutting edges come together cleanly and evenly throughout their length. Try this test with a cheap imported pair of cutters and see the difference! The previous remarks concerning the action of pliers applies just as much to sidecutters. A loose action and the cutting action will be greatly impaired. On the other hand, there is nothing more annoying than a pair of cutters or phers that need two hands to open them!

WIRESTRIPPERS. Different people have different ways of stripping the insulation from wire, varying from the use of the front teeth to an old kitchen knife! If the wire is nicked, and it often is, it will eventually break if it is subject to vibration or flexing. The ordinary sidecutters are frequently pressed into service, as wirestrippers mainly because they will have just been used to cut off a piece of wire and are still in the hand! However there are plenty of wirestrippers on the market designed for the job so there Is no excuse for improvisation.

Wirestrippers usually have some means of adjustment to cater for different size wires. It is important not to try to strip wires of a larger gauge than that for which the wirestripper has been set, or the wire will be nicked. One of the latest style wirestrippers on the market has a four position cam that can be adjusted immediately to another gauge of wire, while still held in the hand. Incidentally, this model also has wire cutters which have ground edges, a great improvement.

A feature on some types of wirestrippers, which the author considers. of great importance, is the provision of a spring that keeps the cutters normally open, the cutting action taking place against the spring.

SCREWDRIVERS. Strictly speaking, every different sized screw needs its own screwdriver, if the blade is to fit the screw slot properly, but not many of us are likely to go to that extreme. The width of the blade should be about the diameter of the screw head or wider, rather than narrower. For electronic construction work it is best to buy a set of screwdrivers with blade widths from about $\frac{1}{8}$ " up to $\frac{1^{\prime \prime}}{4}$ or $\frac{5}{16}$ ". Three screwdrivers should be sufficient. A useful addition is one for the Philips style screws, that is the screws with crossed slots.

Screwdriver sets are available consisting of several blades that can be fitted to a common handle. Although these may be cheaper than individual screwdrivers they are less convenient in practice.

The screwdriver blade itself should be about $6^{\prime \prime}$ or so long depending upon the blade width. One very useful size of blade is one about $\frac{3}{32}{ }^{\prime \prime}$ wide for use with knob setscrews. The blade should be as long as possible so as to reach a knob in spite of the clutter of other knobs, dials etc. on a panel.

Don't take any notice of the markings on some screwdrivers that seem to indicate that one can use them with safety on 10000 volt circuits! Anyone having that sort of voltage around should know the precautions he should take. They would not include the use of a screwdriver on 10000 volts!

Variations on the basic screwdriver include the ratchet

A mosf useful de-soldering aid is a reel of specially treated copper braid which will completely remove unwanted solder from a joint by capillary action.


The Adcola protective shield is intended for wall or bench mounting and has slots for a spindle carrying a reel of solder.

This de-soldering tool (L.S.D. Ltd.) could be invaluable when stripping components from surplus equipment. It is self-contained and simple to operate and may be used as a conventional soldering iron.


The Adamin miniature iron, Model 15, can be obtained complete with a set of four bits of differing size. The tube of special lubricant is used to prevent the bits from sticking to the heating element.


## SIIRLL WORKSHOP TOOLS



Three of the Bib range of wirecutter/ strippers by Multicore Solders Ltd. The latest, centre, is the Model 3 A with an instantly adjustable cam enabling the strippers to deal with most wire sizes. The cutters have ground cutting edges. The ends of the arms are, in fact, flat spanners for 4 and 6BA nuts, a most unusual feature.


A set of Stanley 'Pozidrive' screwdrivers enabling Philips type screws to be reached in the most awkward corners.


The Stanley range of 'Spee-D.Grip' screwdrivers incorporating a spring clip which holds screws in position for starting in inaccessible places.
types which remove the necessity of continually lifting the screwdriver from a screwhead thus, incidentally, reducing the risk of the blade slipping and doing damage. Of course, if one used a screwdriver properly, the blade should never leave the slot, but this seldom happens!

A most valuable addition to a set of screwdrivers is one having a spring attachment at the tip of the blade to firmly hold a screw while it is being inserted into place. Its value in awkward spots is inestimable. Such an attachment can sometimes be bought separately for use on almost any screwdriver.

SPANNERS. There is always a temptation to use a pair of pliers on any nut that needs tightening, usually because the pliers are ready to hand. It should be remembered that the opposite faces of a nut are parallel to each other where-as the jaws of a pair of pliers are only parallel when they are closed. That's why the pliers always slip off the nut and cause damage.

First of all, get a set of box spanners covering 0 to 8BA either single or double ended. The single ended ones with a wooden or plastic handle are recommended. Just two, for 4 and 6BA, are worth their weight in gold in a workshop. Having long stems they are invaluable for placing nuts into odd corners where fingers or pliers cannot reach.

It is always good practice, wherever possible, to tighten a nut on to a screw rather than use a screwdriver on the screwhead. This is particularly important where the screwhead is visible such as on the front panel of a piece of equipment where a damaged slot on a screw would be an eyesore.

Another requirement in the spanner line is a set of flat spanners, again covering the BA sizes, possibly 0 to $8 B A$. These can very well be double ended ones since the length of the spanners is not of much importance.

Some sets are single ended but bolted together at one end so that one can be selected and the others swung out of the way, rather like a set of feeler gauges.


DRILLS. In the past a hand drill would have been a first acquisition in a workshop, followed by an electric drill when the money was available. Today the cost of electric drills is such as to justify buying one from the outset.

The simple electric drill will have a single speed and accommodate drills up to $t^{\prime \prime}$ " in diameter, ('twist drill' is really the proper name for the bit that does the cutting). This model will perform just about all the routine drilling jobs in a workship where electronic projects predominate. A chuck adaptor to take drills up to $\frac{3^{\prime \prime}}{3^{\prime \prime}}$ in diameter is highly desirable, enabling the usual holes for mounting potentiometers etc. to be drilled without having to resort to flles or reamers.

If the drill is likely to be used for other duties around the house then the two speed drill with a chuck taking drills up to $\frac{1}{2}$ " in diameter can be considered. A popular accessory for the electric drill is the speed controller which is an electronic device enabling the speed to be varied continuously from a few r.p.m. to maximum speed. Such controllers are available in kit form or ready assembled.

If a lot of drilling is anticipated, particularly on chassis work, a drill stand will prove a boon. Most electric drill manufacturers have such stands but their design is such that they will accommodate most makes of drill. Such
stands ensure that holes are drilled vertically, which may not be very important on sheet metal work but could be critical with a deep hole. Holes will be more accurate both in size and positioning when drilled on a stand rather than with a hand held electric drill.

When using drill adaptors enabling, say, a $z^{\prime \prime \prime}$ drill to be used in a chuck of nominal $\frac{1^{\prime \prime}}{4}$ capacity; remember not to overload the motor by exerting excessive pressure on the drill especially in hard materials.

Hand operated drills suitable for the small workshop range from the simplest single speed ones with a $t^{\prime \prime}$ capacity to those taking drills up to $\frac{1^{\prime \prime}}{2}$ and having an integral gearbox providing either a low speed or a high speed at the chuck. A little extra spent on a good hand drill will be a worthwhile investment. In particular the gears will run more smoothly and require less energy and the chuck will be of a better quality.
Now, what about the twist drills themselves? These are bought singly, as required, or, if starting more or less from scratch, then in a set complete with a stand for use on top of the bench. They may be had in sizes from about $\frac{1}{32}{ }^{\prime \prime}$ to $\frac{1^{\prime \prime}}{4}$ in diameter or in a numbered range from No. 80 ( $0.0135^{\prime \prime}$ ) to No. $1\left(0.228^{\prime \prime}\right)$. Letter drills carry the range on upwards from ' $A$ ' $\left(0.234^{\prime \prime}\right)$ to ' $Z$ ' $\left(0.413^{\prime \prime}\right)$.

For many years the author has had just four drills always to hand, No's 12, 24, 32 and 43. These will cover the drilling of holes for tapping or clearing for 2,4 and 68 A screws.

As far as the quality of twist drills is concerned, it is necessary to take a little more care over their selection than with other tools since there are a lot of drills on the market that seem to be made of little more than soft iron! So once again beware of cheap imitations and if the pocket will stand it, buy a small set of genuine high speed drills. These will last for ever and seldom need sharpening if not abused.

PUNCHES. Whenever it is required to drill a hole in a chassis, or anywhere else for that matter, it is essential to use a centre punch to mark the spot where the hole is required. Otherwise it is almost certain that the drill will wander off the spot, scratching or damaging the surface being drilled. This particularly applies to hard surfaces and when using a hand held electric drill especially if care is not taken to keep the drill vertical.

There's not much to say about centre punches. A good one will be of tempered steel with a sharp round point, usually machine ground. Spring loaded punches are available which obviate the need for a hammer, so these punches are very convenient and the extra cost well justified.

Chassis punches are a necessity although the number bought for a small workshop will probably depend upon the funds available. These punches will cut holes ranging from $z^{\prime \prime}$ to $3^{\prime \prime}$ in diameter but it is recommended that at least two be obtained, 髫" diameter for B7G valveholders and $\frac{3^{\prime \prime}}{4}$ diameter for B9A valveholders. These punches require a suitable Allen key to operate them and their cutting limit is about $16 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. mild steel. They will cut through the usual $16 \mathrm{s.w.g}$. aluminium sheet like the proverbial knife through butter, thus saving a great deal of time and energy.
A usefultip . . . the resulting hole will have a burr on one side so place the cutting part of the punch on the side of the panel or chassis that might be of some importance from the point of view of appearance.

Other chassis cutters are available that will cut square or rectangular shaped holes but the infrequent need for such cutters does not justify their purchase.
The predecessor of the chassis cutter was the tank cutter. This had an adjustable cutting bar for different diameters of hole, but unless used in a lathe or drill press,


This bench clamp enables a portable electric drill to be used with grinding, polishing or other attachments (Wolf). The drill is a Wolf Sapphire 73.


For precision drilling a bench drill stand is essential. The spring loaded press magnifies the hand pressure some ten times. (Wolf)


The old-fashioned hand drill now has a streamlined appearance. In this Stanley drill the gears are completely enclosed protecting them from dirt, swarf etc. and obviating nipped fingers!


SMPLL UORHSHOP TOOLS


Requiring only a $\frac{7}{10}$ hole to begin operations this Adel nibbling tool will cut a hole of any shape or size in sheet steel up to 18 gauge steel or aluminium up to 16 gauge. (West Hyide Developments Ltd.) The punch and die action cuts without distorting the metal.


This precision made Mole 'Supercut' tool has a powerful action capable of cutting long straight lengths, curves or angles in sheet materials such as steel, aluminium, plastic laminates etc. A replacement blade can be fitted very easily.


The standard Mole wrench has many uses in the small workshop. In spite of its immense grlpping power it can be opened at a touch on the release lever.
the resulting hole could be a little rough in appearance. A slow cutting speed was essential and for this a carpenter's brace was ideal.

Other hand tools for cutting holes in sheet metal include the family of "nibblers'. A small hole is first drilled in the sheet metal, the nibbler inserted and the action of the tool nibbles the material away to form a hole of almost any shape desired. The hole can be finished off with the usual files.

Shears, or tinsnips, will be needed for cutting sheet metal and a $10^{\prime \prime}$ pair with straight jaws should suffice for general work. The will make a straight cut in sheet metal for a limited distance although the 'goosebill' type will usually prove better for this work.

TAPS. Now and again the necessity will arise to tap a thread in a hole and for this two taps for each size thread are required, together with a tap wrench. The first tap is tapered, to start the thread, and the second tap is full size. Sometimes a third tap of intermediate taper is used but this is not really required for small threads.

The tap sizes most likely to be required are the usual 2, 4 and 6BA. Taps can be broken off relatively easily and for this reason a proper tap wrench is strongly recommended.

The simplest and probably the best tap wrench is the bar type with a screw grip on to the tap. The chuck model is more expensive but it does enable the tap to reach positions that would be inaccessible with the bar type wrench.


FILES. For finishing off the edges of metalwork and suchlike a $10^{\prime \prime}$ flat file will be quite adequate. That known as 'second cut' will prove best being neither too coarse nor too fine for work on soft metals. A similar sized half round file is very useful for finishing off large holes such as those required for meters. While smaller holes, for valveholders etc., can be cut with chassis punches the cost of similar punches for large meter holes tends to be prohibitive, especially as they are used relatively infrequently. Such large holes can be made by drilling a circle of small holes, then knocking out the disc of metal and finishing the work with a half round file.
$A \frac{1}{4}$ " round file completes the essential complement of files. This will find most use in enlarging small holes in the absence of a reamer or for 'drifting' holes that are not quite where they ought to be! Perhaps the spot was not properly spotted with a centre punch before drilling!

With the advent into common use of slide switches on electronic equipment the need has arisen for a couple of small needle files, one round and the other square, for finishing off the rectangular holes required in panels for these switches. If a set of needle files can be managed then so much the better as these sets contain files of circular, triangular and flat sections as well as those already recommended.

When buying files get suitable wooden handles for them at the same time. Don't try to use a file without a handle or there will be a grave risk of the tang of the file entering the wrist if the file should jam or stick in the work.

HAMMERS. As far as hammers are concerned the small workshop should have one with a plastic or rubber head for bending sheet metal, without causing undue marking of the metal. A second hammer should be a light
one of the 'ball-pein' variety for general work around the place. A good make is advisable to ensure that the hammerhead is properly attached to the shaft as a hammerhead in orbit around the workshop is not funny!

REAMERS. During the construction of a piece of electronic equipment one always seems to need to drill a hole that is between the sizes of the drills available. In such a case a tapered reamer can be invaluable, as an alternative to filing out the hole with a round file, a process that does not endear one to one's family, if they are within earshot!

A pair of reamers can deal with holes between $\frac{1^{\prime \prime}}{t^{\prime}}$ and $1^{\prime \prime}$ in diameter producing a clean circular hole with little or no burr. They are usually intended for use in a hand brace but can on occasion be held in the hand.

VICES. If a bench or table is permanently available for constructional work then a vice of some sort will prove to be a great boon especially when fabricating metal work. While smaller sizes are obtainable, an engineer's vice with $4^{\prime \prime}$ wide jaws should be the smallest size to be considered. Such a vice is extremely rugged and will take a lot of punishment, including using it as an anvil!

Since the vice jaws are made of hardened steel, and serrated at that, it is essential to make up a pair of soft jaws to protect work held in the vice. These jaws can be made from pieces of aluminium or other soft metal. Other vices have clip-on jaws which are plain, serrated or slotted to hold tubing or round material more securely. Another model of vice has a separate mounting plate that can be attached to almost any flat surface by means of a suction pad.

A carpenter's vice with $6^{\prime \prime}$ or $8^{\prime \prime}$ jaws is even better for sheet metal work since the jaws can be opened to a far greater extent than those of an engineer's vice. However the space needed to permanently mount a carpenter's vice, especially below and behind the vice, is considerably more than that needed for an engineer's vice. However, there are some portable vices on the market that seem to combine the best points of both types.

Although one may possess a bench vice there is always a need for a portable clamping or holding device and this is where the wrench comes in very handy.

The modern version of the wrench has an easily operated lever that allows the wrench to be locked on to the work and just as easily released. A clamp to hold the wrench is also available, which can be fitted to a table or bench top with a single clamping screw, thus providing a fixed vice if necessary. The jaw opening of the wrench can be adjusted over a wide range by a single screw fitting.

RULES. For marking out on sheet metal a rule and a square (trysquare) will be required, if an accurate job is to be done. A $12^{\prime \prime}$ rule is good enough and although it would seem that choosing such a simple tool would be no problem, the choice is, in fact, very wide indeed, ranging from rather inaccurate wooden rules, with printed graduations, to engraved steel ones.

Although a $6^{\prime \prime}$ square can be bought separately it really pays to buy a combination square since this incorporates a rule and square, usually of reasonable quality.

HACKSAWS. The hacksaw will come in for quite a bit of use in the small workshop, mainly for cutting sheet metal to size before forming into a chassis. Some of the cheaper hacksaw frames will be found to twist when the blade is tightened up, so a good quality frame will be a sound buy. The adjusting nuts, too, can be difficult to turn owing to poorly formed threads.


The versatility of the Mole wrench can be greatly increased by the use of this table or bench clamp, converting the wrench into a portable vice.


This Stanley aluminium portable vice has $5^{\prime \prime}$ jaws and can be clamped to any convenient surface. it is eminently suitable for bending sheet metal into chassis etc.


The Mini-Mole is just $5^{\prime \prime}$ long and very useful for getting into those tight corners, inaccessible to its 'big brother'. The curved jaws provide maximum contact with round or irregular shapes.


## smPIL WORKSHOP TOOLS



This set of Philips trimming tools and adaptors will cover every conceivable requirement when aligning electronic equipment. The compact packaging makes the set particularly useful to the service engineer. (Combined Electronic Services Ltd.)


If you don't feel like tramping around the tool shops to gather together your collection of small workshop tools then this very comprehensive tool kit may be the answer. It has a wide variety of spanners, screwdrivers, pliers, cutters, a soldering iron and solder, and even an angled mirror for viewing components tucked away in tight corners. (C.E.S Ltd.)

A frame to take $10^{\prime \prime}$ or $12^{\prime \prime}$ blades is big enough and the blades themselves can be of the low tungsten flexible type. These will deal with general cutting of mild steel, brass and aluminium etc. and absorb a lot of misuse without breaking. For maximum performance high speed flexible blades are to be preferred although they are a little more expensive.

For maximum efficiency the size of the teeth on the blade must be related to the material being cut. For soft metals, such as aluminium, a coarse blade is desirable to prevent the teeth from clogging, but on harder metals such as brass or on thin sheet metal'a finer toothed blade is required. The number of teeth per inch of blade normally runs from 14 to 32 for hand held saws.

A very useful addition to the standard hacksaw is the 'junior' or bow-frame saw in which the blade is held under tension by the bow. The standard blade is $6^{\prime \prime}$ Iong with fine teeth. This saw is very handy for the smaller, finer jobs where the standard hacksaw would prove clumsy. In the electronic workshop the bow saw will be used a great deal for shortening the spindles of potentiometers, capacitors etc. Remember to hold the unwanted end of the spindle in the vice while cutting, rather than the component itself.

TRIMMING TOOLS. Although perhaps it is wandering into the realms of servicing it is not inappropriate in this review to mention trimming tools for electronic equipment. After all, having built a lovely radio set in our nice new workshop, it won't be of much use unless the tools to align it are to hand.

These trimming tools can be bought individually, as needed, but preferably they ought to be obtained as a klt. This will prevent the use of ordinary screwdrivers for adjusting iron cores which can only cause damage in the long run.

A kit may consist of one basic tool holder. with several types of trimming tool that can be fitted into the holder or, In the more expensive kits, a complete trimmer for each purpose, such as adjusting iron cores, capacitors and special types of capacitors, such' as the Philips pre-set beehive capacitors, which are met with quite frequently.

In conclusion it should be pointed out that for every basic tool mentioned in this review there are a dozen or more variations on the market. Like women, there are short ones, tall ones, thin ones and fat ones! Each one to his own fancy! The average D.I.Y. or tool shop is sated with gadgets that promise to relieve us of the tedium and work required to do a particular job. Few succeed, and those that do probably take twice as long to set up to do the job as one would take using basic tools alone.

It's surprising how often one resorts to basic methods and tools in the workshop. If funds are limited then put them to good use by buying only the best in tools from reliable manufacturers. If their names happen to be a household words they would not have reached those dizzy heights by selling rubbish.


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## SPECIAL SERIES-READ IN CONJUCTION WITH OUR FREE WALL CHART!

WHEN the idea for this series was tirsi broached, everything seemed straightforward and simple. We would entitle the wallchart Building Bricks,' said the Editor, in an inspired moment, and the articles would take up each 'brick' and develop it so that newcomers to transistor circuitry could understand what was happening inside those inscrutable black boxes.

Which was all very fine-but it completely overlooked the fact that any beginner worth his salt is going to say 'But Why???'

No good my declaring '. . . an input signal to the base of Tr 1 turns it on, which causes $\operatorname{Tr} 2$ to turn off and $\operatorname{Tr} 3$ and $\operatorname{Tr} 4$ to return to the stable mode, releasing $\operatorname{Tr} 5$, which now bottoms and allows . . 'etc.' It would sound like a paragraph from Henry!

To understand transistor action, it is imperative that we begin by understanding what makes the darned things tick, That means going back to pure physics, not too deeply, preferably without mathematics, and then progressing to the methods of construction and the reasons why there are essential parameters-rules of operation, if you like-and this is what we shall begin with.

## Semiconduction

As the name implies, the semiconductor is neither pure insulator nor electrical conductor, but something in between. Its resistivity lies roughly midway between that of conductors and insulators. This resistivity, moreover, decreases with increase of temperature (the resistance of a conductor increases slightly as the temperature is raised). This is one important difference between semiconductors and conductors.

Next, perhaps more important, is the way in which current flows in a semiconductor. There can be two processes; either a flow of negative electrons, as in the case of a conductor, or a flow in the opposite direction of positive 'holes'. The concept of hole flow is so vital to the understanding of transistor operation that we must begin with a simple discussion of atomic structure in order to get a grasp of it.

## Atomic Structure

An atom consists of a central nucleus containing positively charged protons with a number of negatively charged electrons orbiting around it. The negative charge carried by an electron is equal to, and exactly opposite to, the positive charge of a proton. In a neutral atom, the numbers of each are equal.

Different elements have atoms containing different numbers of electrons, orbiting at different distances from the nucleus. The electrons close to the nucleus are tightly bound to the nucleus, but those circling in the outer orbits are more loosely bound, and are easier to dislodge from the parent atom. The outer ring is called the valence ring, and the valence electrons are those with which we are mainly concerned in this work. The looser they are bound to the atom-in any material-the better a conductor a piece of that material will make. This is because loosely bound valence electrons are free to be attracted by a positive charge, such as a voltage applied across a conductor by means of a battery.

Electrons flow out of the conductor into the battery at the positive connection, but an equal number pass from the negative connection of the battery into the conductor at the negative terminal, keeping the overall total of electrons the same

Conversely, an insulator is a material in which all electrons are tightly bound to the nucleus, so that a large amount of energy is needed to break any loose to make a current flow.

## Crystal Structure

The structure of a crystal of germanium or silicon is depicted in Fig. 1. It can be seen that there are four valence electrons in the outer ring and these are balanced in the nucleus by four positively charged protons, as indicated. In the three-dimensional structure that exists in reality, every atom is equi-distant from four other atoms and each valence electron forms a pair with one from an adjacent atom.

In two-dimensional form this can be shown as in Fig. 1, with pairs of electrons forming co-valent


Fig. 1 : lllustrating the structure of a germanium or silicon crystal.
bonds. This is a perfect crystal, with no impurities; and no electrons are dislodged from their atoms. This is, in fact, the basic structure of a perfect insulator, at absolute zero temperature, but as temperature increases, the energy of each atom is increased and some electrons can break away. If a voltage is applied, these free electrons are attracted to the positive pole and a current flow results. The space left in the crystal structure by a dislodged electron is termed a 'hole' and this represents a net positive charge.

We can therefore think of a movement of negative electrons toward the positive voltage terminal and a virtual movement of positive 'holes' (as successive electrons move to fill the vacant spaces) toward the negative pole.


Fig. 2: Adding a 'donor' atom producing an excess free electron. N-type crystal.

## $N$ - and P-Type Semiconductor

This is a simplified explanation, but suffices to show the principle of hole movement. In a pure semiconductor there must be one hole for each free electron. Such a pure semiconductor is termed an intrinsic material. However, if an element of impurity is added to the intrinsic material an imbalance is obtained. If, for example, a "donor atom" having five valence electrons is added to the crystal, as in Fig. 2, there is an excess free electron. Four of the donor atom's valence electrons form co-valent bonds, in the manner previously described, but the free atom cannot do so and orbits loosely

Some materials with a valency of five are antimony, arsenic and phosphorus. These can be added to the germanium or silicon crystal in proportions of as little as one part in a hundred million and reduce the resistivity of the germanium from $70 \mathrm{ohms} / \mathrm{cm}$ to $1 \mathrm{ohm} / \mathrm{cm}$. Because the donor atoms in such a material contribute free electrons, which carry a negative electric charge, a material "doped" in this


Fig. 3: In this case, an 'acceptor' atom creates a 'hole'. P-type crystal,
way is called an n-type semiconductor (i.e. it possesses an excess of negative current carriers).

The opposite process takes place if indium, borium or aluminium atoms are added to the basic semiconductor material, germanium or silicon. These materials have three valence electrons, and the addition of them creates holes in the semiconductor material, as shown in Fig. 3. The resultant semiconductor material has an excess of positively charged holes and is termed p-type.

## Minority Carriers

Minority carriers are another important factor of semiconductor physics. If a few electrons are injected into p-type material they are called minority carriers (the majority of current carriers in p-type material being, as we have seen, holes). Recombination takes place as the electrons move to fill the holes.

Similarly, holes injected into an n-type material
form minority carriers in this type of material, and small currents arise as recombination takes place. The time this takes to happen, termed the minority carrier lifetime, is an important property of a semiconductor material and the operation of most transistors depends on the movement of minority carriers.

## PN Junction

A single piece of p-type or n-type semiconductor material is purely resistive: reversing the applied voltage does not alter the current flow. But if p-type and n-type regions are formed in a piece of semiconductor material, with the regions adjoining each other, a rectifying device is formed: when voltage is applied in one direction, current flows, but in the other direction little or no current flows. Such an arrangement is termed a pn junction.

In Fig. 4, the region to the right of the centre line is n-type, possessing a certain density of free electrons. To the left of the line is a region of p-type material possessing a number of positively charged holes. When the junction is first formed, it becomes an area of activity for a short period as free electrons tend to cross into the p-type region and holes move over to the n-type region.


Fig. 4: Showing the charges existing in an unbiased PN junction.

This process of diffusion results in a potential building up across the junction, as the p-region gains a net negative charge with respect to the $n$ region in the vicinity of the junction, and vice-versa, and this potential prevents further diffusion. This potential is termed the barrier potential and the region in the immediate vicinity of the junction, where there are very few free current carriers remaining after the initial diffusion, is termed the depletion layer or region, as shown in Fig. 4.


Fig. 5(a) : Position of the charges with forward biasing and Fig. 5(b): with reverse biasing.

## Forward and Reverse Biasing

We now have a piece of material with pronounced non-linear characteristics: when an external voltage is applied, current will flow in one direction only. If the p-type side is made positive with respect to the n-type (see Fig. 5(a), forward biasing), the effect is that the barrier potential is reduced. Majority carrier holes will diffuse across to the n-type (where they become minority carriers), and majority carrier electrons will move in to the p-type region. A relatively large current can flow across the junction in this way.

Reversing the bias, i.e. applying it as shown in Fig. 5(b), increases the barrier potential (since the external voltage adds to that across the depletion layer already) so that current flow is made very low -in fact virtually nil. However, due to the effect of temperature a number of holes and free electrons will be generated on either side of the junction-this happens even at room temperatures-and this will give the effect of a small reverse current flowing across the junction. The forward current may be quoted in milliamps but the reverse current only in microamps.

## Characteristics of PN Junction

Fig. 6 shows the characteristics of a pn junction. As the forward voltage is increased from zero, the forward current increases slowly at first until the junction barrier is overcome, then rises rapidly. (In alternative terms, it can be said that the resistance of the forward biased junction is low.) In the reverse direction, the current remains low and practically


Fig. 6: The electrical characteristics of a PN junction.
constant as voltage is increased, i.e. reverse resistance is very high. Eventually, however, a point is reached where the reverse voltage accelerates the reverse current carriers to such a speed that they break down covalent bonds in the crystal structure, thereby releasing more carriers and producing an avalanche breakdown (when the junction may be destroyed). The value of voltage at which this occurs depends on the resistivity of the material.

As the reverse saturation current is due to the effect of heat, it increases with temperature, giving

# 1.4 <br> E: © 

SOME months ago these notes described the Motorola MC1596 balanced mixer integrated circuit and outlined its operation as a multiplemode detector in receiver circuits. The unit discussed now goes further; full i.f. amplification facilities, at any chosen frequency up to 12 MHz ., are incorporated, as well as multiple-mode detection. It is the National Semiconductors (U.S.A.) type LM373, available in the U.K. from: Athena Semiconductor Marketing Co. Ltd., 140 High Street, Egham, Surrey: or Rastra Electronics Ltd., 275 King Street, Hammersmith, London W.6.

## Uses

The unit is intended, not just to operate in each mode, but to make it possible to design a receiver in which intermode switching would be simple without reduction of performance. As well as the standard a.m. and f.m. modes, as used for broadcast entertainment in the m.f. and v.h.f. bands, the LM373


Fig. 1: The equivalent circuit of the LM373 (below) with (above) the top view of the pin connections to the various 'blocks'.



Fig. 2: External circuitry required for the LM373 when used as an AM IF strip.
can handle n.b.f.m. (narrow-band frequency modulation) increasingly used in mobile radiotelephone systems, and s.s.b. (single sideband suppressed carrier) transmissions of commercial or amateur origin.

## Design

Considerable ingenuity is exercised in the design of the i.c. to reconcile the conflicting demands of each transmission system with the minimum of intermode switching in the receiver. A.M. and s.s.b. receivers must have a linear amplitude response, maintained, despite the wide range of signal strength received off the air, by an automatic gain control subcircuit. An f.m. i.f. amplifier strip, on the other hand, is essentially nonlinear, in that any input signal is amplified to a limiting level before application to the detector stage. It follows that automatic gain controlyas no place in such a system.

To provide the facilities indicated, the LM373 incorporates four gain blocks, which can be operated in a limiting mode, as well as an automatic gain


Fig. 3: The LM373 in its role of an FM IF strip suitable for narrow or wideband reception.
control system and two detectors, one a balanced mixer similar to that in the Motorola circuit already mentioned, and the other a peak detector for a.m. operation. The equivalent circuit and pin diagram given in Fig. 1 allows these sections to be identified.

## Operation

Some discussion of the circuit would appear necessary. Each gain block is essentially the familiar differential amplifier, which can be operated in its linear region when the circuit is used for a.m. or s.s.b. reception, or driven to full gain as a limiter for f.m.

The automatic gain control characteristic is secured through a pair of transistors shunting the differential interstage coupling components. When no a.g.c. voltage is applied, these transistors are cut off, but as the control voltage is increased, they draw off an increasing fraction of the signal current; finally the point is reached where the signal from


Fig. 4: The SSB/CW IF strip shown here further demonstrates the versatility of the LM373.
the first gain block is shorted out, and coupling to the second is merely through stray capacitance.

This approach secures the unusual a.g.c. range of over 70 dB ., enabling linear a.m. detection to be attained with input signals from 40 microvolts up to 200 millivolts. The string of reference diodes and emitter follower transistors from the d.c. +ve. line provide power supplies with decoupling to each stage of the amplifier strip, together with appropriate bias voltages to set the operating point of each element. As well as simplifying the external decoupling requirements of the circuit, these imply that each stage essentially has its own regulated supply, so minimising the effects of power line variations.

The balanced mixer stage is easily identified by the cross-coupling of its component transistors. For f.m. it operates as a quadrature detector, a system outlined when describing the MC1496, as was the product detector system used for s.s.b. In a.m. operation, this section is deliberately unbalanced by an externally applied bias, leaving only one side of
the differential system connected to its load effectively.
N.b.f.m. demodulation requires a quadrature element of a higher $Q$ factor, i.e. more rapid change of phase with frequency, and therefore a quartz crystal is substituted for the more familiar l.c. circuit. The final section is the peak detector, acting as a.m. detector and a.g.c. voltage source for linear a.m. and s.s.b. operation.

A few precautions are necessary for successful operation. The a.g.c. stage output at pin 9 has a very low impedance, being taken from an emitter follower. A value of 70 ohms is quoted, so that if a high $Q$ crystal filter is employed here a series resistor is advocated, so that the resonance peak of the component will not be damped. The manufacturer of the crystal in question will quote a recommended source impedance, allowing selection of an appropiate resistor.


Fig. 5 : In TV circultry the LM373 can be used as an audio/video amplifier with a.g.c., manual gain control or gating.
The d.c. feedback loop at pin 3 must be carefully bypassed, since the decoupling must be effective throughout the passband of the LM373; an electrolytic or tantalum capacitor, as used for a.f. decoupling, has too high a reactance at high frequencies to be effective alone, so a small disc ceramic or silver mica is added.

For s.s.b., bypassing of r.f. at pin 7 is particularly important, as any local oscillator component reaching the peak detector would increase the a.g.c. voltage generated, and lead to an unnecessary reduction of system gain. The optional nulling circuits may be used to maximise gain should that prove necessary (an unlikely event).

Figs. $2-5$ indicate the circuitry in which the LM373 may be employed, and it should be noted that the communications receiver applications for which it was designed do not exhaust the potential of this versatile unit, as is indicated by the circuit for a wideband amplifier indicated in Fig.5. Other reported applications have included use as an amplitude-modulated r.f. oscillator, an s.s.b. generator, or indeed any of the applications open to the balanced mixer circuit in which extra wideband gain is useful. So presumably more will be heard of the LM373!

## TRANSISTOR CIRCUITRY FOR BEGINNERS

-continued from page 509

the effect shown by the dotted line in Fig. 6.
Junction devices of this type are known as diodes, and widely used in radio, television and other electronic applications. The diode is essentially a one-way valve, as we have seen from the forward and reverse bias characteristics.

Germanium diodes were the first types developed for commercial use, but the current-handling capacity of these devices was limited, and it was not until research provided a means of treating silicon that power rectifier semiconductors became a practical proposition. Silicon is more difficult to melt and purify than germanium, but can be doped with donor and acceptor impurities in the same way as germanium, is less sensitive to temperature, and will handle higher powers more efficiently. Common types pass currents of half-an-ampere and more with a practically negligible forward resistance. Silicon diodes have a high reverse voltage rating. (On the other hand, a germanium has greater efficiency at low voltages and at higher frequencies, and is thus retained for signal circuits.)

The pn junction forms the basis of most forms of semiconductor diode and transistor. One other form of construction is important in connection with diodes-the point contact type. In this the rectifying property is possessed by a contact between a thin metal wire and a piece of semi-conductor material. Though junction devices predominate today, point contact diodes are still much used as detectors in radio equipment. Their characteristic is similar to that of the pn junction, shown in Fig. 6, except that as the reverse current increases beyond the reverse breakdown point, reverse voltage tends to fall.

TO BE CONTINUED

## TELEVISION

## SEPTEMBER

## TV INTEGRATED CIRCUITS

Integrated circuits are beginning to appear in the new TV chassis in increasing numbers, particularly in colour receivers. The types used have been specifically developed to fulfil the signal processing requirements of TV sets but nevertheless bring with them numerous new techniques. In a new series starting this month we shall be examining these new i.c.s. to familiarise readers with the various changes in TV set design they involve.

## SERVICING TV RECEIVERS

Recent articles have concentrated on fairly up-to-date chassis. This month however we are going back to a widely distributed " 625 -line convertible" chassis, the Sobell ST195/ST282 series. This will give us an opportunity to examine the faults common in older sets.

PLUS ALL THE REGULAR FEATURES

## SERVIONG 

## PART 6

## G. J. KING

## THIS MONTH G. J. KING RETURNS TO THE SERIES TO ELABORATE ON AUDIO MATTERS, DEALING WITH CONVENTIONAL VALVE AMPLIFIERS AS WELL AS TRANSISTORISED HI-Fi EQUIPMENT.

NOW that colleague Hellyer has cleared the decks on alignment, superhet theory and f.m. stereo, it is my turn to have a go on audio equipment.

## GRAM AUDIO STAGES

An ordinary radiogram differs little at the preaudio stages from a radio receiver, so much of the information presented earlier will also be applicable here. However, most contemporary radiograms are equipped for f.m reception, possibly with a decoder to abstract the stereo information and to direct the left and right channel signals separately to the corresponding audio channels. Thus, many radiograms and record reproducers feature a pair of audio channels and speaker systems. These are very much twin partners, the design of one following very closely the design of the other.

A mono radiogram has just the single audio channel and speaker system (though some designs employ more than one speaker unit, the mono signal being applied to them all in suitable frequency division). Whether mono or stereo, the f.m. front-end is a fairly common factor, the demodulator these days usually being the ratio detector in unbalanced or (in better quality equipment) balanced mode.

However, while the f.m. detector in a mono receiver feeds the audio channel direct (via the $50 \mu \mathrm{~S}$ deemphasis), the f.m. detector in a stereo receiver delivers the multiplex signal to the stereo decoder, the function of which is to process the sidebands of the stereo information (the difference signal) in conjunction with the mono information and the 38 kHz reference signal (derived by frequency multiplication from the 19 kHz pilot tone) to retrieve the left and right transmitted audio signals in the best isolation (e.g., with the least crosstalk).

De-emphasis and sometimes subchannel suppression (deleting the residual 19 kHz and 38 kHz multiplex components which can interfere with the audio section and add whistles to tape recordings) are engineered into the left and right audio channels, prior to the power amplifiers.

I shall be having something to say about f.m. front-ends, i.f. channels and stereo decoders later, but for now let us concentrate on the audio channels.

The simplest audio channel is designed round the triode-pentode valve, the triode receiving the detector's signal, amplifying it and then passing it to the control grid of the pentode for power amplification, Fig.1. Quite a lot of valve radiograms are equipped with this kind of audio section, as also are the inexpensive species of record reproducers. These were in fashion when stereo started, so early stereo models boast two such audio sections.
The advent of stereo was concurrent with the development of transistor equipment, which was fortunate for the receiver designers, so the audio channels we now find in stereo eqipument are mostly transistorised.


Fig. 1: A typical mono audio amplifier using a single triode-pentode valve,

Some designers favoured the use of a triode voltage amplifier (often one section of a double-triode valve) followed by a separate power pentode. This allows for greater audio power than generally possible from the pentode section of a triode-pentode valve, particularly so far as the early triode-pentode valves are concerned. Such a circuit is shown in Fig. 2 , which will be seen to have much in common with that in Fig. 1.

Better quality receivers have been made with a pair of triode-pentodes in each channel to provide push-pull working. In this type of circuit the two pentodes are the push-pull output valves, while the two triodes are arranged as phase splitters supplying the anti-phase drives to the pentodes. However, since few receivers like this are in the field, there is not much point in detailing the circuit, but it is worth bearing in mind.

Let us return to Fig. 1. The audio signal from pickup or radio detector is coupled by Cl to the top of the volume control VR1, the selected level of signal then being applied via C 2 to the triode grid. Note that the grid is returned to chassis through R1 ( $10 \mathrm{M} \Omega$ ) and that the cathode of the triode is returned direct to chassis. This is a common trick to secure grid bias for the triode. Contact potential in the valve causes a minute current to flow through RI, and because of the high value of this resistor a small potential developed across it, and since it is negative at the grid it serves as grid bias. It requires an electronic testmeter to measure this bias (since a normal meter would swamp the $10 \mathrm{M} \Omega$ with its own resistance).

A common cause of distortion in this kind of circuit stems from a faulty triode, the fault resulting in excessive grid current and hence a potential across RI in opposite polarity to that of the bias. The triode anode current rises, and often the valve tends to overheat.

R2 is the triode anode load, with R3 and C3 the decoupling components. The signal across $\mathbf{R 2}$ is coupled by C4 to gl of the pentode, which is biased by the voltage drop across the cathode resistor R4, the grid being returned to chassis through R5. C5 is the cathode bypass capacitor (always an electrolytic in audio output stages).

The anode of the pentode is loaded by the low impedance speaker via the output transformer $T 1$. C6 across the primary attentuates high-order har--monic distortion and makes the reproduction more palatable, as there is no negative feedback and no cancellation of second-harmonic distortion by any push-pull action.

## FAULTS

Owing to the simplicity of this circuit, (Fig. 1) faults are not generally very difficult to define or locate. A cause of bad distortion is poor insulation of the coupling capacitor C4. This would permit a positive potential to exist at the control grid of the pentode with a consequent counteracting of the negative bias. It would be accompanied by excessive positive voltage at the cathode of the pentode, which could result in the failure of $C 5$. The valve would also


Fig. 2: A development of Fig. 1 with a separate triode amplifier. The other half of V1 drives the output valve in the second channel.

If one channel only is in trouble one can easily compare the voltages, etc. with those at the same points in the active channel, which is a quick way of bringing the fault to light. It is also possible to substitute the same type of valves (V2, for example) between the two channels, which would tell conclusively whether the valve is responsible.

There are refinements in Fig. 2, not found in Fig. 1 circuit. For example, voltage negative feedback is applied from the secondary of the output transformer T1 to the cathode of the voltage amplifier V1, and for this to happen one side of the secondary has to be connected to chassis. The amount of feedback is determined by R1 and R2, while C1 in parallel with R1 corrects the phase of the feedback signal, ensuring that it remains substantially negative over the audio spectrum. Thus, if there is a significant rise in gain (e.g., full output at a low setting of the volume control VR1) and abnormally high distortion, then R1 and R2 should be checked for value.

C1 open-circuit could encourage instability at certain frequencies, and this is sometimes 'triggered' by transients of the music. The oscillation could be at a frequency above audio, and the only way to prove this is by connecting an oscilloscope across the secondary of the output transformer T1, preferably when the amplifier is driven by a sinewave signal from an audio generator. The oscillogram in Fig. 3 gives some idea of what the spurious signal could look like on the sinewave. The audio generator would be coupled to the grid of the triode, of course.


Fig. 3: This oscillogram shows spurious oscillation triggered off by transients in the signal

If the earth and feedback connections are reversed on the transformer the feedback would be positive and the amplifier would turn into a violent oscillator! Take care, therefore, when fitting a replacement transformer! Moreover, designs like this are geared to an output transformer of specific characteristics, so if it becomes necessary to replace the transformer, and the exact type is not available, then it might be necessary to alter the feedback resistors and the phase-correcting capacitor so that the values provide the correct conditions of feedback.
An interesting feature of the circuit is the tap on T1 primary. An output transformer of this kind serves both for speaker coupling and for 'smoothing' the h.t. supply fed to the earlier stages. In fact, it acts in part rather like a smoothing choke. In the
circuit, C 2 is the reservoir capacitor across the power supply output and C3 is the main smoothing electrolytic.

R3 and C4 series combination across the signal part of T1 primary attenuates the higher-frequency distortion components while, with C5, providing a degree of corrective phase shift.
The triode V1 in this circuit is cathode-biased, the main cathode resistor being R4, decoupled by C6. VR2 is the stereo balance control, which feeds either side the 100 -ohm resistors (one to the cathode circuit of the partnering channel). The 100 -ohm resistors provide a small degree of negative current feedback since they are out of the decoupling influence of C6. The voltage developed either side of VR2 relative to the slider also contributes to the feedback. Thus it will be appreciated that by adjusting VR2 from its centre setting, the feedback on one channel will decrease, while increasing on the other channel. This, therefore, allows the gains of the two channels to be adjusted relatively for the most desirable stereo effect.

VR3 is a simple 'tone' control which works as a reactive potential-divider in conjunction with C7. With increasing frequency the reactance of C 7 falls, which means that as the slider of VR3 is turned towards this end of the track the bass is lifted and the treble diminished.

It will also be seen that two speaker units are employed. That for bass is connected directly across the secondary of $T 1$, while the treble unit receives drive via R5 and C7. R5 reduces the power, while C7 passes only the higher frequency signals. This is a very simple frequency-divider, which cannot be regarded as 'hi-fi.'

## HI-FI CIRCUIT

One channel of a slightly different circuit which, possibly differing in detail, is found in some hi-fi equipment. This is the power amplifier department only, Fig. 4, where $\operatorname{Tr} 1$ is the input transistor, coupling to the d.c.-coupled pair $\operatorname{Tr} 2 / \mathrm{Tr} 3$ via C 1 . The coupling is from Trl emitter (the collector 'earthed' to signal through C2). The stage is thus an emitterfollower, which has a relatively high base input impedance, giving the correct match from the control section.
$\operatorname{Tr} 4 / \operatorname{Tr} 5$ constitute a complementary driver pair (complementary because one is n.p.n and the other p.n.p.). These drive the push-pull output transistors $\operatorname{Tr} 6 / \operatorname{Tr} 7$ alternatively into conduction and cut-off on each half-cycle of signal. It will be noted that the base of $\operatorname{Tr} 4$ is fed from the same source (e.g., $\operatorname{Tr} 3$ collector) as the base of Tr 5 . Thus on a positive halfcycle $\operatorname{Tr} 4$ conducts and pushes a positive half-cycle into $\operatorname{Tr} 6$ base. At the same time $\operatorname{Tr} 5$ is driven downwards and since $\operatorname{Tr} 7$ base is connected this time to $\operatorname{Tr} 5$ collector a negative half-cycle appears at Tr7 base, pushing this transistor into cut-off. On a negative half-cycle of drive the conditions reverse, so that $\operatorname{Tr} 6$ is driven into cut-off and $\operatorname{Tr} 7$ into conduction.

The two half-cycles are reconstituted as a complete waveform in the load comprising the speaker, coupled through C3.

C4 is a bootstrapping capacitor which applies feedback in such a way as to increase the driver input impedance. A.C. negative voltage feedback is applied to Tr2 emitter circuit from the push-pull pair, via C5

Fig. 4: Circuit of one channel of a transistorised hi-fl stereo amplifier.
and R1. Feedback of 60 dB is a common hi-fi _value, which reduces the output impedance to a fraction of one ohm.
D.C. coupling exists right through to the output transistors from Tr2, so by adjusting the d.c. conditions, both the drive 'balancing' and the quiescent current of the output transistors are affected. Some circuits have no means of adjusting these parameters, so if the value of a critical component alters, so might the quiescent current and balancing.

However, in Fig. 4 VRI adjusts the balancing and VR2 the quiescent current, which leads us neatly to hi-fi equipment fault-finding and tests. But before going on to this it is noteworthy that temperature stabilisation is often provided in or around VR2 circuit to hold the quiescent current steady with increase in temperature of the transistors. The 'thermal element,' TH1 in Fig. 4 provides this function.

## AMPLIFIER TESTS AND FAULT FINDING

For detailed testing in audio circuits, especially high quality circuits in which transistors are employed, we require an audio signal generator, audio millivoltmeter and an oscilloscope, at least. For distortion tests, we require in addition a distortion factor meter or similar instrument.
If it is assumed that the radio or pickup source signal is reaching the input of the amplifier and there is no output from the speaker, then we should check as far as possible into the amplifier with a signal tracer (e.g., pair of headphones, etc.-see Part 3, PW July 1971). In Fig. 4, for example, we could get to the base of $\operatorname{Tr} 2$, but possibly not any further owing to the d.c. cbuplings of the subsequent stages. Indeed, it is often necessary to consider the power amplifier proper as a complete circuit from the testing point of view. This is because a faulty transistor or component in one stage could reflect fault conditions into the other stages as a result of the d.c. couplings. This possibility is often overlooked when testing in circuits of this kind.

When we reach the point where the signal disappears, therefore, we may have to continue the exercise in terms of voltage measurements at the electrodes of the various transistors, comparing them with the correct values given in the service manual or on the circuit. Complete failure commonly implies significant change in the d.c. conditions, so the fault area should soon be revealed by voltage testing. 'If a coupling capacitor has failed, then signal tracing should reveal this without much trouble.
More difficult faults to locate are those resulting in low gain, intermittent crackling and distortion at all powers or only at high power. Starting first with distortion, a preliminary test should be that of

quiescent current (assuming the amplifier to be 'Class B,' which it usually is in these days of transistors).

If possible, one should avoid connecting a milliammeter in series with the output pair for this test, owing to the resistance of the meter movement. In a circuit like that in Fig 4, the voltage across the emitter resistor (R2) of $\operatorname{Tr} 7$ should be measured on a d.c. millivoltmeter. The voltage will be very low because R2 is only 1 ohm ! Anyway, suppose we measure 25 mV , then by Ohm's law be can calculate that the current is 25 mA ( 1 ohm resistor, since $I=V / R$, where $I$ is in $m A$ when $V$ is in $m V$. A current of 25 mA is, in fact, a common quiescent current value, and it can be adjusted accurately by a preset such as VR2.

If there is no preset and the current is way out, the values of the resistors associated with this part of the circuit should be measured (also the thermistor, etc., if used). Excessive quiescent current will cause the output transistors to run hot and limit the output


Fig. 5: Equipment set-up required to check the balance of output transistors.
power, while too low a value will precipitate high distortion (crossover distortion) at low volume.

Balancing requires an oscilloscope to monitor the output sinewave fed in from an audio generator. The setup is shown in Fig. 5 where is will be seen that the speaker is replaced by a load resistor of suitable power rating (to match the steady-state power of the amplifier). The volume control should be fully advanced, and the generator signal increased until the tips of the displayed waveform commence to clip. When the balancing is correct both tips will clip together. If one clips before the other then the balancing preset (such as VR1 in Fig. 4) should be adjusted for symmetrical clipping. The waveforms are shown in Fig. 6.


Fig. 6(a) Symmetrical clipping of the waveform when the gain is advanced indicates correct balance.

Fig. 6(b) Output waveform with an unbalanced stage.


We can use the setup in Fig. 5 to measure the steady-state power of the amplifier. For this test the oscilloscope should either be accurately calibrated in peak-to-peak voltage (so that the r.m.s. voltage of the waveform can be determined by multiplying the peak voltage-i.e., half $p-p$ voltage-by $0 \cdot 7$ ) or an r.m.s.reading audio millivoltmeter should also be connected across the load, as shown in Fig. 5.

The load voltage just prior to the clipping point should be noted, and the steady-state power in watts is equal to the load voltage squared divided by the load value in ohms (i.e., $W=V^{2} / R$ ). The power is usually taken at 1 kHz , but the power bandwidth can be appraised by noting the reduction in power at the low and high frequency ends of the spectrum.

Low power, but good waveshape often signifies low supply rail voltage, and this can result from low mains voltage or incorrect mains voltage tapping.

## DISTORTION TEST

Finally, total harmonic distortion testing. For this we require either a complete distortion factor meter or the set-up in Fig. 5 with a distortion measuring unit which, as was noted in the Test Instrument supplement, April 1971, is a 'notch filter' which
tunes out the fundamental frequency of the output signal, leaving only the harmonic components, the voltage sum of which is compared in ratio with the voltage of the fundamental frequency across the load, to give the dB or percentage total harmonic distortion.

The complete set-up is shown in Fig. 7. The voltage across the load is first measured at the required power (use the oscilloscope for maximum power test to indicate the clipping level) with the switch in postion $A$, then switching to position $B$, allowing the audio millivoltmeter to read the voltage of the summed harmonics. For this to happen, however, the distortion measuring unit needs to be very accurately tuned and balanced for the least indication on the audio millivoltmeter, which requires the range of the millivoltmeter progressively to be stepped down. The total harmonic distortion is the voltage ratio between the readings at switch postions $A$ and $B$.


For example, if voltage $A$ is 9 V (which across 8 ohms would correspond to about 10 W ) and voltage $B$ is 90 mV , the ratio would be $100: 1$ or 40 dB , corresponding to $1 \%$ t.h.d. With 9 mV at B it would represent a ratio of $1000: 1$ or 60 dB , corresponding to $0 \cdot 1 \%$ t.h.d.

Remember, though, that a simple distortion measuring set 'notches out' only the fundamental, leaving wideband noise as well as t.h.d. components, so the distortion readout will include noise. Residual hum is generally removed by a high-pass filter in the test set-up, but such filtering, of course, cannot be used for distortion measurements at low-frequency for obvious reasons.

## END OF PART SIX

## Back Numbers Important Announcement

We regret to inform readers that owing to the closure by the Company of the department concerned it will no longer be possible to supply back numbers of Practical Wireless and Television.

To ensure obtaining regular copies of these magazines readers are strongly urged to place a regular order with their local newsagent, or to take out an annual postal subscription.

Reference to past issues of the magazines may sometimes be obtained at certain public libraries who may hold bound volumes. A few libraries are said to offer a photostat service. Alternatively, we are always willing to insert a free request for specific back numbers in our "CQ" column which appears in most issues.

# practically $\underset{\substack{\text { Wirless } \\ \text { comentaver br } \\ \text { HENRY }}}{ }$ 

NO, not that Ted. Henry is not taking the lid off political chicanery-leave that to out-of-office politicians. TED, or, more properly, T.E.D., is a beautiful phrase $I$ heard during a broadcast interview recently. That ubiquitous character, A. Spokesman, was defending the Post Office in the face ; of polite but relentless attacks from the BBC, who were following up listeners' complaints about crossed lines.
'How does it happen,' came the suave question, 'that we pick up our telephone and find our conversation interrupted by a pair of drooling lovers or an irate customer who has lost his laundry?'
Ah, well, our indefatigable spokesman countered, that is probably caused by a Transient Equipment Defect. Isn't that lovely? Imagine, Joe, when wifey grumbles about the birdies that chirp unwantedly during her 'Housewives' Choice', being able to say airily: "Can't do much about that, I'm afraid. That's TED."
The point is that the faults are not just short-lived, ephemeral, unidentified phenomena-giving them a name has fixed them, preserved them for posterity, entered them on the list of recognisable bugbears.

Wireless work is full of such


We cheer when we get a half-scale reading ...
instances; for example, the intermittently blowing fuse. We pursue diligent tests before replacing the offending glass tube. We measure resistance, cheer when we get a half-scale reading, then remember that we intended to change the battery of the meter last time we used it.

Baffled by the lack of shortcircuits we decide to beat the brute by checking the current. Meter in series-ah, which way round? Switch on. No, the other way around: is that pointer bent? But the reading is no greater than expected, and certainly less than the rating of the fuse, which we refit. Perhaps it is the result of overheating. We cover the now smug-looking set with our jacket and wait.

It is a bad habit of some manufacturers to let dirty great heatsinks jut out at the back, and the fallacy about cool-running transistors is shattered as we smell singeing Harris Tweed and realise we have a Class A output to contend with.
Touching the fins to confirm our nose, we snatch fingers back with a howl and, sucking them, muse that the problem could not have been temperature at all. And the fuse still holds.
Obviously, TED has been at work. He is only waiting for us to settle comfortably with a book and a glass at our elbow before blowing the fuse again. Or, more likely, lulling us into a false sense of security, will allow us the whole evening's listening, then, while we sleep through the small CQ hours, will sneakily contract that vulnerable thread of wire again so that rising betimes, we gaily throw the switch and get: silence!
TED is no fool. That Post Office bloke urged 'sweet reasonableness' upon subscribers when their lines were crossed. The inference was that if we pretended not to notice the fault it would go away. Any $P W$ reader could have told him that is the wrong approach.

. . . and we snatch our fingers back with a how/.

How many times have you had a piece of equipment propped up on the bench, festooned with test leads, bedizened with warning lights, burbling away in the corner of your mind, and leaped into the breach as the expected fault occurs, only to be greeted with a return to normality the moment you touch the chassis?

Tracking down TED is the work of an expert. One tests piecemeal. One uses not sweet reasonableness, but bitter reason. Diagnosis, as our soon-to-finish series on Servicing has emphasised, is the secret of succéss. And for diagnosis, where TED is concerned, read inspired guesswork. This is where LID comes in. LID is one of those contractions I am no longer likely to forget, since talking to Bob Stockwell, publicity manager of Bang \& Olufsen, the day after an exhausting exhibition.
'How do you manage to organise it?'
'It's not easy,' he said, 'but we begin by catering for the Largest Imaginable Disaster and take it from there. . .,

That is the secret of dealing with TED. Put the LID on him: when the fuse blows, imagine the worst that could happen and assume that it has! But don't try to telephone the repairman-you will probably get the laundry company again.


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## THE BROADCAST BANDS Malcolm Conah

## Frequencies in kHz - Times in GMT

THE first report this month comes from a new reporter D. A. Hairon from Jersey in the Channel Islands. The equipment used was a Codar CR70A receiver, a 50ft long wire and a 19 metre band dipole. This combination enabled him to hear the following stations:-
9480 Radio Peking in English at 0130.
9625 CBC , Canada in English at 0200.
11710 R. Nacional de Espana in Spanish at 0130.
11720 CBC, Canada in English at 0200.
11730 R. Nederland, Bonaire, English at 0200.
11745 HCJB, Quito, Ecuador, English at 0000.
11775 TWR, Bonaire in Spanish at 0215.
11810 RAI, Rome in English at 0100.
15195 R. Ankara, Turkey, Turkish at 1600.
15250 RSA, South Africa in French at 1856.
15300 HCJB, Quito, Ecuador in Spanish at 2145, and in Portuguese at 0000.
15305 SBC, Switzerland in English at 1315.
15345 Radio Kuwait in English at 1630.
15355 ABC, Australia in English at 0530.
17655 R. Cairo, U.A.R. in English at 1730.
17715 ABC, Australia in French at 2315.
17810 R. Nederland, Bonaire in Spanish at 2130.
17830 R. Lebanon in English at 1830.
17825 NHK, Japan in English at 0800.
21525 WNYW, U.S.A. in English at 1800.
21640 R. Nederland, Bonaire, English at 1830.
25730 R. Norway in English at 1200.
Ray Warren of Bury St. Edmunds has used his Koyo 1661 receiver to good effect and heard the following:-
3980 Voice of America in English at 1720.
5970 R. Free Europe, Germany at 1450.
6075 WNBC, New York, news at 1630.
6085 R. Nederland in English at 0930.
9030 R. Peking in English at 2050.
9390 R. Tirana, Albania in English at 1530.
9750 SBC, Switzerland, news in English, 1540.
15300 HCJB, Quito, Ecuador, English news, 1410.
Peter Hall of Pocklington, Yorkshire, has a Meridian 10 transistor superhet and a 50 foot aerial which enabled him to hear:-
6070 R. Sofia, Bulgaria in English at 1930.
6130 HCJB, Quito, Ecuador in English at 0800.
7275 RAI, Rome in English at 1940.
9550 Radio Finland in English at 1815.
9580 BBC , Ascension Island relay at 2115.
9620 R. Belgrade, Yugoslavia at 1800.
9625 Israel B.C. in English at 2045.
9630 R. Trans Europe, Portugal at 2100.
9700 R. Sofia, Bulgaria in English at 2140.
11620 AIR, Delhi in English at 2010.
11672 R. Pakistan in English at 2030.

11755 R. Finland in English at 0930.
11915 BBC, Cyprus relay at 1645.
11965 TWR, Monaco in English at 1700.
Ross Pullen of Crawley used his Murphy A72 communications receiver and a 100 foot long-wire to hear some interesting stations:-
6015 VOA, Rhodes, music and talk at 2200.
6145 R. Nacional, Brazil in Spanish at 2245.
9510 R. Noumea, New Caledonia, French at 0910.
9605 HCJB, Quito, Ecuador, English at 0730.
9745 R. Baghdad, Iraq in English at 2015.
9775 R. Diffusion-Television Nationale Congolaise, African music at 2200.
11672 R. Pakistan, English at 2002.
11965 Rwanda relay of Deutsche Welle at 2037.
17705 R. Havana, Cuba, Spanish news at 2015.
F. Wall of Walthamstow, London, E.17, has a homebrew 5 -valve superhet and a 20 foot wire aerial, this equipment enabling him to hear:-
6035 R. Warsaw with Polish news at 1620.
6070 R. Sofia, Bulgaria in English at 1930.
9912 AIR, Delhi English news at 2000.
9630 R. Sweden 'Calling DXers' at 1230.
9660 R. Trans Europe in English at 1250.
9833 R. Budapest, Hungary, requests at 1945.
11735 R. Belgrade, Yugoslavia, English at 1530.
11740 ABC, Australia at 1530.
11755 R. Finland in English at 1500.
11765 ABC, Australia, sports programme at 0830.
11955 BBC, Malaysian relay at 1815.
15325 CBC, Canada, sports news at 1200.
21460 HCJB, Quito, Ecuador sign-on in French at 2015.

21535 RSA, South Africa, English news at 0930.
21595 CBC, Canada with news at 1910.
25790 RSA, South Africa, English news at 1500.
Stephen Mathews of Hull with his Bush 4-valve domestic receiver and 90 foot loft aerial heard the following:-
9660 YVLM Radio Rumbos, Venezuela at 0320.
9833 R. Espana Independiente, clandestine at, 2230. 11765 ZYB8, R. Diffusion, Brazil at 0220.
11795 WINB, Red Lion, U.S.A. English at 2100.
11815 ZYW24, R. Brazil Central at 0045.
15260 FEBA, Seychelles testing at 1700.
17890 BED-40, Taiwan, weak at 0330.
21690 WIBS, Grenada in English at 1945.
21720 R. Ghana in English at 1445.

Reports should arrive by the 15 th of the month and be addressed to the author at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.
 IGH noise levels have been the main complaint this past month with many people confessing to giving up and going off to watch television (sacrilege!). Despite these conditions there was still a large number of listeners who persisted, despite the noise, and managed to log some worthwhile DX as reward for their labours. Just goes to show, while you are busy moaning about the QRM and QRN, someone else is just getting on with it and working half the world.
News that G5ZT has received slow-scan television pictures from FG7XT in Guadeloupe makes you think. Many people would be only too pleased to hear an FG7 on c.w. John Rimmer (Deal) warns that VS9MF is on from Gan-a small island in the Maldive group of islands. Listen on 14 and 21 MHz ; sorry, no specific freqs given.
No activity reported on ten metres this month and it rather looks as though everyone has migrated to 14 MHz . This latter band is an early morning/late night affair nowadays with everything dying out from mid-morning to late afternoon. On fifteen metres the same remarks apply, although occasionally South American stations appear during the daytime. For those who would like to persist on 28 MHz , suggested time for a listen is from 1400 to 1600 hrs .

Incidentally, this is one of the best bands on which to play with aerials since these normally work out to a very reasonable size. A quarter wave vertical is only around the 8 ft . 6 in . mark, exact size depending upon frequency since ten metres is one of the widest bands we have among those l.f. of 144 MHz . If sunspot counts are any indication, then September would be a very good time for ten metres
A few contests this month just to get your ears in trim for the long winter evenings. September 11 and 12, WAE phone contest; 12, $3 \cdot 5 \mathrm{MHz}$ Field Day; October 2 and 3, u.h.f./s.h.f. Field Day; 2 and 3, Oceania contest-a chance to bag a nice list of VK/ZL callsigns; 2 and 3, IARU Region 1 u.h.f./s.h.f. contest. Don't forget the Cumulative Activity on 70 MHz from October 9 to December 30. This is our "use or lose" band and all support is very welcome, if not essential. You might try listening for some of the RAEN activity on four metres. The Radio Amateur Emergency Network has numerous "local" groups and the favourite band is 70 MHz . How about listening out for your local group, or perhaps even joining them? Name and address of the Secretary of RAEN is; E. Bassett, 57 Upper St Helen's Road, Hedge End, Southampton, SO3 4LG.

## Logs

A solitary log for two metres arrived this month from David Sheekay (Ashford). Gear is a superregen using three transistors and an i.c. audio amplifier, homebrew, and antenna is described as "telescopic". Stations logged include; G2JF, G2VB, G3EFX/P, G3JXN; G3LTF, G3TDP/M, G3UNT,

## THE AMATEUR BANDS David Ghbson, G3JDG

## Frequencies in kHz - Times in GMT

G4AAR, G4AAZ/M, G50X, G8BYC, G8CCO, G8EBV, G8ELW and F0UJ. What about it all you converter and superhet types-you can't let a super-regen show you up, or can you?
Richard Mortimore (Cardiff), is now up to 20 w.p.m. at c.w. and says that he recommends all s.w.l.s to learn morse. Small sample from his list of c.w. pickings on 21 MHz includes; CR6ID, JA3IL, KZ5BB, PY2EYE, PY4UK, W4NZK, WB4QVQ.

Mark Marsden (Ilford) uses a cassette tape recorder which has its input wired in parallel with the headphones. He claims this is very useful for checking calls heard against calls logged afterwards. Seems a very good idea. Calls heard, checked, recorded and logged on 14 MHz include signals from; CE5ID, JA3JH, KV4HE, KZ5JF, PY2ZAC, PY4AEB, SV0WZ, VK3ART, VK4PJ, VP2AA, WA1DYR/MM, WB6YIY/MM, YV5CPA, YN2SC, ZL5ISJ, 4X4JU, 4X40C, 9Q5IA.
"My uncle has given me all the PWs he can spare", writes Kevin Davies. Receiver is an R107 and the antenna 120 ft . l.w. The following were in evidence on 14 MHz s.s.b.; CT1ZE, EA3KO, EA4LM, EL2C, EL4A, HR3VSJ, JA1VK, K1WPS/CT1, LU4DMG, W1CMG/P1, W1MYA/MM, WA5KHM, WB6YIY, 4U1ITU, 4X4HT.
Paul Newman is at a new QTH which is some 680ft. a.s.l. Antenna at present is an 80 ft . end fed and the receiver a modified BC348R ex G3HXS. With this gear, and from his lofty perch at Aylesbury, Paul bagged these on 14 MHz ; OD5BV, PY6AT, PY8LM, (Sao Luis Island), WA5STY/P/YV5 (Caracas) 4X4BL, 4X4GV, 5Z4KL, 9K2AL.
Jim Martin says that as it's the start of his holidays he thought he'd send in a report. (May the sun shine on your antenna, Sir.) His recentlypurchased HA500 is nourished with an inverted V and rewards on 14 MHz were; CR6EF, CR7IK, EL2CY, FP8CS, HC6CB, JA3IXL, JA4EDY, JA6FUV, JY9XL, KG4AN, KH6FP, KH6BB (Kure Island), KL7DIV, KR6US, KV4AD, KZ5AO, LG5LG, MP4MBC, PY1FN, PJ2MI, PZ1AD, PZ1AW, TG9GF, VK3JS, VP2MO, VQ9R, VS5CB, VS9MT, VU2BEZ, YV5EC, ZE1DO, ZP5GU.
On $3 \cdot 5 \mathrm{MHz}$ c.w. it was UA1YUZ. Up on 7 MHz , signals came from CX6CG, LX1BJ, PYCVX, PZ1AX, UW3NG, ZB2A, and ZP5AR. Gear is a JR500SE with 132 ft . end fed. Operator's ears belong to Julian Iredale (Llandudno), and up on 14 MHz s.s.b. Julian heard CR6GA, HV3SJ, OG5A, (Finnish Field Day prefix), PY2ZAC, PY4ATL, PZ1DL, TI2J, 5A4DL.
P. McKay advises that he never notices any logs for 7 MHz in the Amateur Bands, and sends in a list of G-stations heard on this band. Well, the G's must have been working someone, so how about a listen this month?

[^3]
## PREMIER HI-FI OFFERS!

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case (List $£ 36$ )
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MANY readers have written to ask us if we can suggest any gramophone records that may help them to bring alive even more vividly the days of vintage radio. After a certain amount of research we have discovered that Fountain Press, the book publishers, have now announced a new "Fountain" record label of Vintage Jazz music. The first 12 in . LP releases feature historic jazz and blues recordings but future plans include Music Hall, Opera and other music of the "Good Old Days".
The records are housed in attractive double sleeves which open, displaying an authentically documented note section with accurate historical background information and photographs.
The first two releases are: "Original Dixieland Jazz Band and the Louisiana Five"-from the extremely rare Aeolian sessions of 1917 (catalogue number FJ-101). "Ladds Black Aces-Vol. 1"-the first 16 sides recorded by another pioneer white jazz band during 1921-1922 (Cat. No. FJ-102). Following soon are "Naylor's Seven Aces" (c 1923-4). "Jelly Roll Morton" $1923-4$ piano solos. "Lovie Austin's Blues Serenaders" (1924-6), and "Annette Hanshawe. Vol. $1^{\text {" }}$-Miss Hanshawe enjoyed public acclaim as a singer of popular songs in the 1920's and 1930's and her records are great collectors' items today.
These records are available from record shops at $£ 2 \cdot 10$ or direct at $£ 2 \cdot 25$ which includes post and packing, from Fountain Records, 46 Chancery Lane, London WC2A 1JU.

If you have any questions or queries on these records or any other Vintage records or performers please write to us at Practical Wireless and we'll do our best to help you.

## $\mathbb{T h e}$ Atystery of the $\mathcal{S} \mathbb{C} .200$

We would, at this point, like to thank the many enthusiasts who have been kind enough to write and help us try to clear up the mystery of the missing John Scott Taggart design "ST.200" mentioned in "Going Back", August 1971.

One theory put forward by Mr. F. W. Snow, Belper, Derby, is that there never was a design called the ST. 200 but one called "The ST Twin"-a 2-valve reflex. It was published in The Wireless Constructor and not the Popular Wireless magazine during the late 1924's and early 1925's.

Mr. J. Crossley hailing from Failsworth, Manchester, also maintains that the ST. 200 never existed. He
says that J.S.T. visited the United States just after the advent of the ST.100. On his return, he was full of enthusiasm about a gentleman by the name of Armstrong, and Popular Wireless'published a circuit by this man which J.S.T. had brought back with himi. It was the "Armstrong Regenerative Circuit". J.S.T. then went on to publish the ST. 300 design, so it appears that Armstrong's receiver may have taken the place of the missing ST.200.
Mr. A. H. Jenkins, Dringhouses, York, writes to say that as far as he recollects, the ST. 200 may well have been just a mains-operated version of the ST. 100 and therefore did not differ operationally from that design. It did however, need an extra valve. Mr. Jenkins also informs us that the S.T. series did not start at ST.100, although this was the most famous. He says that he used to have a limp covered book of J.S.T. circuits which started at crystal sets and worked up through one valvers. He seems to remember an ST. 37 which was quite wellknown at the time.

## Hintage Bets

Mr. William Pozniak, from 37 Manley Road, Oldham, Lancashire, has loaned us some photographs. He says that for three years he has been routing through old cellars and derelict radio shops throughout the North (we hope you came out for food now and again, Mr. Pozniak) and has produced the collection of Vintage gear shown in the first picture. He says that it takes him most of his time to strip down and restore these sets to as near as possible their original condition. The sets he has are as follows: The Burndept Screened Four, The Burndept Portable Five, the Burndept Ethophone Five and the Burndept Ethophone Two.

Mr. Pozniak believes there were about 20 Burndept receivers. This figure is agreed by Mr. Keith Lancaster who we mentioned some time ago as wanting to start a Vintage Radio Club. Incidentally, response has not been too good and if readers are interested in such a Club, would they please contact Keith Lancaster, 40 Great Gardens Road, Hornchurch, Essex.

The rest of Mr. Pozniak's collection contains a Marconiphone Type 32 three valver, a domestic Cossor two valver, a small set in a metal case named a GECoPHONE Victor 3 and a few home-constructed receivers.

The receiver shown in the second picture is a


An overall view of Mr, Pozniak's collection of Burndept receivers. The small receiver shown at the top centre of the picture and the set in the centre of the picture together with frame aerials are shown enlarged below.


The James Super Six All Waver (c 1927).

1927 James Super Six All Waver (medium, long and short). The reader says that he purchased this set from an old radio shop which was a booming concern in the 1920's, together with a BTH horn speaker, and the person who sold it to him, told him, as he removed the set from the dusty shelf, that he was the first person to touch the set in 40 years.

The neat little receiver shown in the third photograph seems to Mr. Pozniak to be an unusual type, for he says that there appear to be two resistors and


Louwe receiver with encapsulated components.
two capacitors encapsulated in glass tubes which have then been assembled inside the glass envelope of the valve. The make on the case reads "Louwe Radio Berlin" and on the valye it states "Louwe Radio British Made". This set is in working order on headphones although it was originally made to drive-some sort of speaker. The reader would like to know the date and origin of this set if anyone can help him and he would also be interested if anyone could supply him with original Burndept valves for any of their makes of receivers.

## TITANIC DISASTER

We would like to thank the many readers who have written to us about the Titanic Disaster. We are gradually answering all the letters and we hope you will bear with us.

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# MIOITMIHLOMER R. F.GRAHAM 

THIS unit has four switch-selected bands, covering from 1600 kHz to 100 MHz and it can be used as a wavemeter, field strength meter or phone transmission monitor. It is fully enclosed, selfcontained, and requires no battery.

Figure 1 shows the circuit. One pole of the switch, S1, selects the inductors L1, L2, L3 or L4, which are tuned by VC1. Pole S2 switches the diode D1 to the appropriate small coupling winding. When the circuit is tuned to resonance by VCl , rectified current is indicated by the meter.

To monitor phone signals, high impedance headphones or earpiece can be plugged into the phone jack and the phones are then in series with the meter. Cl is for r.f. and a.f. by-pass purposes, while C 2 is an r.f. by-pass capacitor.

## INDUCTORS

L1, L2 and L3 are wound on ${ }^{1}{ }_{2}$ in. diameter paxolin formers. These can be prepared by cutting $1^{1}{ }_{2} \mathrm{in}$. lengths of $\mathrm{I}_{2} \mathrm{in}$. diameter tube, and drilling small holes to secure the ends of the windings. The tubes can later be fixed with adhesive, with small brackets or by having them a push fit on small discs of insulating material screwed in place. Surplus or junk box formers could probably be utilised, but if these are of different diameter it may be necessary to modify the number of turns to suit.

L1 is wound with 34 s.w.g. enamelled copper wire. The tuned portion $A$ to $B$ is 65 turns, in a compact pile about ${ }_{16}{ }^{16}$. long. A space of about ${ }^{3} 16 \mathrm{in}$. is left and section $B$ to $C$ is wound on in the same direction, and has 8 turns side by side.

L2 also uses 34 s.w.g. enamelled copper wire. A to $B$ is 50 turns wound side by side. A space is left as before, and a 4 turn coupling winding ends at C.

L3 uses 24 s.w.g. enamelled wire and has $12^{1}{ }_{2}$ turns from $A$ to $B$, and 2 turns from $B$ to $C$. Figure 2 shows the general construction of L1, L2 and L3.

The windings are doped, smeared with clear Bostik 1 or with some other suitable coil cement.

L4 is self-supporting, of $16 \mathrm{~s} . \mathrm{w} . g$. bare wire, and is shown in Fig. 3. It is wound and shaped to have 8 turns in all and is $1_{2}$ in. outside diameter, and lin. long. $B$ is a tapping 2 turns from end $C$. This coil is soldered directly to tags $A$ and $C$ of the 4 -way switch as shown in Fig. 3. A lead runs from tapping $B$


## $\star$ components list

VCT 200 pF Wavemaster variable capacitor.











Fig. 2 : The general construction of coils L1, L2 and L3 wound on tin. diameter paxolin formers. Details of 14 are given in the text and in Fig. 3
directly to the moving plates tag of the variable capacitor.

## CONSTRUCTION

The wavemeter was built to occupy a rigid plastic 'snack box' approximately 6in. $\times 3^{1}{ }_{2} \mathrm{in} . \times{ }^{1}{ }_{2}$ in. deep. A piece of paxolin sheet, a little under $6 \mathrm{in} . \times{ }^{1}{ }_{2} \mathrm{in}$., was cut to fit inside the box lid. The wavemeter is built wholly on this paxolin panel.

Screws holding L1, L2 and L3 are countersunk. A piece of stout paper to bear dial markings is cut the same size as the paxolin, and cemeted to it. The wavemeter is then assembled and calibrated and the scales are marked. The meter and fixing nuts of the switch, jack, and VCl are then removed and the paxolin panel is fitted inside the box lid. This places the scales under the transparent material of the lid. Paxolin and lid are held together by the meter and the nuts mentioned.

Holes for the jack, switch and VCl are first drilled in the paxolin and the meter hole cut with an adjustable washer or tank cutter. The paxolin is then placed in the lid, and the positions of holes marked on the lid itself, so that they can be made to match up correctly with those in the paxolin.

The box should be of material which is reasonably strong. It should be drilled and cut with tools which are really sharp, and without forcing, to avoid cracking.

A suitable case could be assembled from $1_{4} \mathrm{in}$. thick wood, glued and pinned at the joints. The paxolin panel would then be large enough to be fixed with small wood screws.


Fig. 3 : The wiring details of the wavemeter monitor. L4 is shown on the left and is mounted directly on the switch itself.

All wiring is shown in Fig. 3. Connections should be of stout wire and run clear of each other, and should be reasonably short and direct, especially for L3 and L4.

VCI is 200 pF , but this exact value is not essential, provided that the ranges overlap a little. If an alternative diode is fitted, be sure it is suitable for frequencies up to 100 MHz .

A $100 \mu \mathrm{~A}$ meter would be a little more sensitive than the one specified, but would also cost a little more. Should a $500 \mu \mathrm{~A}$ meter be to hand, it is suitable, though less sensitive. The sensitivity with a 1mA meter becomes rather low so an instrument in the neighbourhood of $200 \mu \mathrm{~A}$ is probably most suitable.


An internal yiew of the completed prototype.
The dial cursor is a $2^{3}{ }_{4} \mathrm{in}$. piece of perspex about $5_{8}$ in. wide, marked with a hair line. Readings will be confused if VC1 can rotate through $360^{\circ}$, so the extreme corner of the back moving plate was turned up with pliers to limit rotation to about $180^{\circ}$.

Values used for C1 and C2 are not particularly important. A medium or high impedance pair of headphones will give best general results.

## CALIBRATION

Actual band coverage with the inductors described is as follows

| Range 1 | $1 \cdot 6-4 \mathrm{MHz}$ |
| :--- | :--- |
| Range 2 | $2 \cdot 8-10 \mathrm{MHz}$ |
| Range 3 | $8-32 \mathrm{MHz}$ |
| Range 4 | $21-100 \mathrm{MHz}$ |

This gives 160 and 80 m bands in Range 1, 80 and 40 m bands in Range 2, all h.f. bands $20-10 \mathrm{~m}$ in Range 3 , and $15-10 \mathrm{~m}$ bands and harmonics in Range 4, without having to use the extreme h.f. end of the ranges, where readings become close together.

The simplest means of calibration is to use a grid dip oscillator, placing its coil 2 in . or so from the appropriate inductor in the wavemeter. This allows complete calibration of all ranges with a minimum of difficulty.

A transmitter can be used for calibration in the amateur bands, and these will probably be of most importance. Place the wavemeter some distance from the transmitter, its dummy load or the aerial lead or tuner, as required to obtain a deflection on the meter.

Some signal generators have an r.f. level meter. If the output of the generator is coupled to the wavemeter by a small loop, a dip may be observed on the level meter when the wavemeter is tuned to resonance. This offers another possible way to secure calibration.

If the wavemeter is coupled to a receiver having a sensitive S-meter, a slight dip in the S-meter reading can be observed when the wavemeter is tuned to the receiver frequency. Input to the receiver can be from a signal generator, or transmission.

All calibration is made with no external connection to the wavemeter and minimum coupling to avoid unnecessary inaccuracy.

## USES

A wavemeter is normally used to indicate frequency bands and harmonics, not exact frequencies within a narrow band. It is placed or held to pick up sufficient r.f. energy from the circuit being investiga-


#### Abstract

TRANSISTORS




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| 1844 | 10p | AAZ17 | 12/p | BY100 | 17\% | OA9 | 10 p |
| IS113 | 15p | BA100 | 15 p | BY103 | 22tp | OA47 | 071 p |
| 18120 | 15 p | BA102 | 22tp | BY122 | 4718 | OA70 | $07 \frac{1}{1 p}$ |
| $\underline{15121}$ | $17 \frac{1}{}{ }^{\text {p }}$ | BA110 | 32 p | BY124 | 15p | 0 O73 | 10p |
| IS130 | 12tp | BA115 | ${ }^{072} \mathrm{t}$ P | BY126 | 15p | OA79 | 09p |
| IS131 | 12 ${ }^{\text {p }}$ | BA141 | 3215 | BY127 | 17tp | OA81 | 072p |
| 18132 | 15p | BA142 | 32tp | BY164 | 57\% | OA85 | 0718 |
| ${ }^{15920}$ | ${ }^{071} \mathrm{p}$ | BA144 | 12tp | BYX10 | 22tp | OA90 | 07\% |
| ${ }^{15922}$ | 07 p | BA145 | ${ }^{20 p}$ | BYZ10 | ${ }^{35 \mathrm{p}}$ | OA91 | 076p |
| 15923 | $0^{07}$ 2 ${ }^{2}$ | BA154 | $12 \frac{12}{}{ }^{\text {p }}$ | BYZ11 | $32 \frac{1}{2}$ p | O495 | 072p |
| IS940 | 07\% ${ }^{\text {p }}$ | BAX13 | 12 1 p | BYZ12 | 30p | OA200 | 10p |
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# TAKE 2® JULIAN ANDERSON 

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"WHAT's an automatic light?" you may well ask. Well, lights are only needed when it's dark, and a very simple electronic switch circuit can be made to turn a small light on as soon as the general light level drops below a certain level; this has its most obvious application in motor cars. Many parts of the country have local regulations requiring cars to display lights shortly after sunset, even when parked. In other areas car parking lights are needed only when the local street lighting is turned off, which in some areas is as late as 2 a.m. There cannot be many people who relish the thought of going out to their cars at that hour to turn on the side or parking lights.

Porch lights which automatically turn themselves on after dusk are another application and although the circuit as shown is not really meant for this, many readers will realise the potential and modify the circuit to operate a thyristor or relay to apply the mains supply to the porch light.

## THE CIRCUIT

Like last month we are making use of a light dependent resistor, that is a resistor whose value in ohms falls with an increase in light falling on it. Also like last month this component is coupled into the base bias potential divider circuit, thus controlling the bias on the transistor, Trl. Rl may have to be altered in value in order to make the "switch" operate at the correct light level.
As darkness falls, the potential at the base of Trl falls as the value of the LDR rises and after a certain point Trl begins to turn off, less current is passed and the potential between the emitter and the collector rises. This in turn starts to turn $\operatorname{Tr} 2$ on. This transistor, when light was falling on the LDR, thus holding Trl on, was off since the potential between base and emitter was held at a very low level due to the small potential across Trl. As soon as Tr2 begins to conduct this increases the current through R4, thus causing Trl to cut off even more and in turn switching $\operatorname{Tr} 2$ more on. This action takes place very quickly and $\operatorname{Tr} 2$ is switched fully on, applying the supply voltage across the bulb causing it to light up.
This switching circuit is known as the Schmitt Trigger and a very useful circuit it is too, with applications in many fields of electronics. As can be seen, only a few components are required and the switching action, using modern transistors, is very rapid and very positive. As soon as the potential at the base of Tr either reaches or falls to a preset level, $\operatorname{Tr} 2$ will switch.
The supply necessary to operate the circuit is taken directly from the car battery, though it is good practice to take this via the car's fuse as any


Fig. 1: The circuit of the automatic light with the lead connections of the transistors shown on the right.


Fig. 2: A suggested component layout on a small piece of tag board.

## $\star$ components list



TAKE 20-continued from previous page
mistakes or serious component failure could create a fire risk.

Car batteries are capable of giving very high currents and therefore the quiescent current is of no great importance but it is in fact low, in the order of 2 mA . The recommended bulb draws 40 mA and it can be seen that these currents can be supplied by a number of dry batteries, though operation in this manner will not be cheap.

The bulb should last a considerable time as it is being slightly under-run; there is of course a small volts drop across R4 and Tr2.

## CONSTRUCTION

A suggested layout is shown in Fig. 2 but this is not critical. One point must be watched however; if the light from the bulb is allowed to fall onto the LDR it will switch the circuit in the same way that sunlight will. This will cause the circuit to oscillate; this can easily be seen as the bulb will flicker or flash. To avoid this the LDR must be pointed away from the bulb.

## WAVEMETER MONITOR-by R. F. GRAHAM

continued from page 530
ted. Tuning resonance is shown by maximum meter reading. In driver and similar circuits with small power, the wavemeter needs to be reasonably close to the circuit inductor. For P.A. and high power circuits, it can be at some small distance.

To check the frequency of low level multipliers and similar circuits, or those which are inaccessible, a link line may be used. This is twin flex or co-axial cable about 2 ft . long with a loop of two or three turns each end. One loop is held near the equipment inductor, and the other near the wavemeter inductor. Clipping a lead to jack and aerial socket also allows the coil primaries to be used for coupling. Resonance will be shown by a dip in grid current in a subsequent stage.

To monitor the audio of an outgoing transmission, plug in phones, tune to resonance and situate the unit where pick-up gives adequate volume.

To tune an aerial, transmitter, tuner or coupling device for maximum radiation from the aerial, a vertical rod or wire is plugged into the aerial socket of the unit, which should be at some distance. Adjustments are then for maximum meter reading, which corresponds to maximum field strength. It is also possible to link couple a dipole or aerial-earth system to the wavemeter for this purpose, or take a short aerial to VCl , ignoring calibration.

The wavemeter produces no signal. It can be used to find bands on a t.r.f. receiver by adjusting the latter until oscillation just arises, and holding the wavemeter coil in line with the receiver coil. Resonance when tuning the wavemeter is then shown by the receiver going out of oscillation.

## BLUEPRINT SERVICE

We would like to draw readers' attention to the fact that the BLUEPRINT SERVICE has been discontinued and therefore no further BLUEPRINTS are available.


I$N$ the beginning there were photocells enclosed in glass envelopes. Variations of these were more sensitive and multi-electrode secondary emission cells appeared like the 931A photomultiplier tube.

In more recent years, many experimenters, and indeed, most of the electronics industry too, has opted for solid state when thinking in terms of a light sensing device. An example here is the ORP12 light-dependent resistor, not to mention the earlier plastic-encapsulated transistors with some of their paint removed!

In America, work is currently proceeding on a technique known as charge coupling for imaging devices. One oxidised surface of a tiny silicon chip is covered with some 288 minute electrodes. These are arranged into groups of three, the last electrode in each group being a common conductor for the whole array.
The area immediately beneath each group of three elements forms a tiny light sensitive device in its own right. Minority carriers are generated in the silicon beneath the electrodes whenever light is focussed onto them. The greater the intensity of the light, the greater the charge, thus forming a direct relationship between light intensity and electrical charge.
The centre electrode of each group is purposely made more positive than the other two elements and thus the charges on minority carriers are attracted to, and congregate at, these centre electrodes.

Individual charges are next moved along the array from one electrode group to the next by putting a successively greater positive voltage on the electrode next to the one which is holding the charge. Eventually all the charges will have been passed along to the end of the device where they are collected by an output electrode.

Because the size of the charge will be dependent upon the amount of original light applied, the final electrode has, in effect, an analogue electrical signal which has a linear relationship to original light patterns of intensity.
How big is a $100,000 \mu \mathrm{~F}$ capacitor-think of the size of some electrolytics. Now imagine how big a one Farad capacitor would be. Just to stretch the imagination even further, think of a 50 Farad capacitor. Fantastic-but they exist and are for sale. Twist in the tale is that the 50 F is contained within one-third of a cubic inch. The devices will also retain their storage (if charged with a voltage) and leakage is so low that even after 16 months they will have retained over 97 per cent of the original charge. Think; they could be charged up and used as a form of battery. When discharged, they could simply be recharged again ad infinitum.


USED EXTENEIVELY BY INDUSTRY，GOVERNMENT DEPARTMENTS －Low cost

OTHER RANGES TO ORDER 200 RANEES IN ETOCK

| NEW＂SEW＂DESIGNS！ |  |  |  |
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| CLEAR PLASTIC PETERS MEAKELTE MANE |  |  |  |
|  |  |  |  |
|  |  | TYPE S－80 |  |
|  | $100 \times 80 \mathrm{~mm}$ ． | 80 mm ． |  |
|  |  | quare fronts |  |
|  | 30V．D．C．．． 58.971 | 50 5 A $\ldots \ldots$ ． 8 8－18t |  |
| $\begin{array}{lll}60 \mu \mathrm{~A} & \ldots . . & 88.471 \\ 50-0-50 \mu \mathrm{~A} & \\ 88.87!\end{array}$ |  | 50－0－50\％A 82.974 | 50V．D．C |
| $100 \mu \mathrm{~A}$ ．．．．．新－871 | 1 amp ．D．C． 88.8 |  | 300 V ．D．C． |
| 100－0－100 10.488 .25 | 5 amp. D．C．${ }^{\text {a }}$ S 971 | 500 2 A ．．．${ }^{\text {¢ }}$ | $\underline{\mathbf{1}} \mathbf{\text { amp．D．C．}}$ |
|  | 300V．A．C．A8．97 |  | $300 \mathrm{~V} . \mathrm{A.C}$ ． |
| 1 mA ．．．．．． 49.971 | VU Meter． 38.75 | 20V．D．C．．粚47t | VU Meter．． |

＂SEW＂CLEAR PLASTIC METERS

Type mR．85P．4tin．$\times 4$ 4iz．fronta．


4 in．$\times 4$ itin．fronts．

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Trpe ED．107．Size overail 100 mm $90 \mathrm{~mm} \times 108 \mathrm{~mm}$ ． A new range of high quality moving coll for school experi－ ments and other bench applications．
$3^{\prime \prime}$ mirror acale．The emonstrate internal working．Arailable in the following ranges：


${ }_{60 \mu \mathrm{~A}}^{25 \mu}$

| $60 \mu \mathrm{~A}$ |
| :--- |
| $50-0$. |
| 100 S |

$100 \mu \mathrm{~A}-100 \mathrm{MA}$ ${ }_{1 \mathrm{raA}}^{500 \mathrm{~A}}$. $1-0-1 \mathrm{~mA}$.
$\stackrel{1}{5 \mathrm{~m}}$
10 mA
50 mA
100 mA


EDGWISE METERS
T7po PE．F0．${ }^{8} 17 / 38 \mathrm{gin}, \times 1 \quad 15 / 3 \mathrm{in}$ ．$\times$


MULTIMETERS for EUGRY purposq

$\begin{array}{lr}\text { TRCE } & \text { PT－34．} 1,000 \\ 0 . P, V & 0 / 10 / 50 / 250 /\end{array}$ $500 / 1,000 \mathrm{~V} . \mathrm{A} \cdot \mathrm{C}$ and D．C $0 / 1 / 100 / 500 \mathrm{~mA}$.
D．C． $0 / 100 \mathrm{~K} . ~ \$ 1-97 \mathrm{t}$. P．\＆P． 12 k p．


> MODEL TEE－200 20，000 O．P．V Mirror scale，overioad protec tion．0／5／25／125／1，000 V．D．C． 0／10／50／250／1，000 V．A．C．0／50 | $\mathrm{uA} / 250 \mathrm{~mA} .0 / 60 \mathrm{~K} / 6 \mathrm{meg} \Omega$ |
| :--- |
| -20 to +62 db. |
| 8.75. | P．\＆P．15p．

MODFH TVE－70． 30,000
O．P．V． $0 / 3 / 15 / 60 / 30,0000 / 600 /$
1，200v．D．C．0／6／30／120／600／
$\begin{array}{ll}1,200 \mathrm{v} . & \text { A．C．} \quad 0 / 30 \mathrm{uA} / 3 / 30 / \\ 300 \mathrm{~mA}, & 0 / 16 \mathrm{~K} / 160 \mathrm{~K} / \mathbf{1 . 6 M} / 16\end{array}$


MODKSL TR－12 20，000 O．P．V．0／0，6／6／30／120 600／1，200／3，000／6，000v．DC $0 / 6 / 30 / 120 / 600 / 1,200 \mathrm{v}$ ．A．C．
$0 / 60 \mu \mathrm{~A} / 6 / 60 / 600 \mathrm{~mA}$ $0 / 60 \mu \mathrm{~A} / 6 / 60 / 600 \mathrm{~mA} .0 / 6 \mathrm{~K} /$


## MODEL PL486

 $20 \mathrm{k} \Omega / \mathrm{Volt}$ D．C． 8kの／VoltMirror BCale．
$.6 / 3 / 12 / 30 / 120 / 600 \mathrm{~V}$ D．C． $3 / 30 / 120 / 600 \mathrm{~V}$ A．C． $50 / 600 \mathrm{ua} / 60$
fon $\mathrm{mA} .10 / 100 \mathrm{~K}$



MODEL 50080,000 O．P．F． with overioad protection． $100 / 250 / 500 / 1,000 \mathrm{v}$ ． $1 / 2 \cdot 5 / 10 / 25 / 100 / 250 / 500$ ． $1,000 \mathrm{v}$ A．C． $0 / 50 \mu \mathrm{~A} / 5 / 50 /$ 000mA． 12 amp．D．C． 0／60K／6 Meg．／60
8．87t．
Meg $\Omega$ ． 28－871．Post paid


THE MODEL TW－80CB FEA TURES RESETTABLG OVER－ LOAD BUTTON．Eensitivity： $20 \mathrm{~K} \Omega / \mathrm{Folt} \quad \mathrm{DC} .5 \mathrm{~K} \Omega / \mathrm{Volt}$ A．C．D．C．Volts：0－0．5，2．5， 10 ， $60,260,1,000 \mathrm{~V} . \mathrm{A} . \mathrm{C}$ ．Volts：
$0-2.5,10,50,250,1.000 \mathrm{~V}$ $0-2.5,10,50,250,1,000 \mathrm{~V}$. D．C．
Currents： $0-0.05,0.5,5,50$ Currents： $0-0.05, \quad 0.5,5,50$,
500 mA ．Resistance： $0-5 \mathrm{~K}, 50 \mathrm{~K}$, $0-500 \mathrm{~K} .5 \mathrm{MEG} \Omega$ ．Decibels： 020 to $+52 \mathrm{db} . \$ 11.50$ ．P．\＆ P ． 17 \＆p．

 100.000 O．P．V． 6 fin．
Scale Buzzer Short Cir－ cuit Check．Senaitivity： 100，000 O．P．V．D．C．5K ．$, 2.5,10,50,250,1,000$ 5，2．5，10， $50,250,1,000$
V．A．C．Volts： $3,10,50$, 50， $250,500,1,000 \mathrm{~V}$ ． $10,100,500 \mathrm{~mA}, 2.5,10$ ${ }^{\text {arop．Resistance }} \mathbf{1 0} \mathrm{K}, 10 \mathrm{~K}, 100 \mathrm{~K}, 10 \mathrm{MEG}$ ， $100 \mathrm{MEG} \Omega$ ．Decibela：-10 to 49 db ．Pias．
 $x 61 \mathrm{in} . x$ 3itin． 218.80 ．P．\＆P． 25 p ．

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1 MA HETERS
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MW1－6 60 mm square
MW1－8 $80 \mathrm{~mm} \begin{gathered}\text { s3．87\％} \\ \text { square } \\ \text { s4．97t }\end{gathered}$
P. \& P. extra

TE－40 HIGH SENSITIVITY A．C．VOLTMETER 10 meg input 10 ranges： $1 / \cdot 003 \cdot 1 / \cdot 3 / 1 / 3 / 10 / 30 / 100 /$ $300 V$ ．R．M．s． $4 \mathrm{cps}-1-.2 \mathrm{Mc} / \mathrm{s}$.
Decibels -40 to +50 dB. Supplied brand new complete with leads and instructions． Operation 230 V, A．C．A17－50． mer． 25 p ．

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with ${ }^{3}$ and lnatrument 1－6－1，500v．A．C．volts $1 \cdot 5-1,500 \mathrm{v}$ ．A．C．Volts Vols up to 1,000 megohms． 200／240v．A．C．operation． Complete with probe and instructions
817.50 ．P．
Additional probes arail．



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in meter protection $0 / 3 / 12 / 60$ in meter protection 0／3／12／60 120／300／600／1，200 $0 / 6 / 30 / 120 / 300 / 600$ $0 / 2 \mathrm{~K} / 200 \mathrm{~K} / 2 \mathrm{M} / 200 \mathrm{M}$（12 Amp． $0 / 2 \mathrm{~K} / 200 \mathrm{~K} / 2 \mathrm{M} / 200 \mathrm{M} \Omega$
to +17 bB. 173p．

TE－900 20，000 $2 /$ VOLT GIANT MULTIMETER， Mirror scale and overload
protection． 6 in．full view meter． 2 colour scale． $0 / 2.5 /$ 0／25／12．5／10／50／250／1，000i 5,000
$100 / 500 \mathrm{~mA} / 10$ $02 \mathrm{~K} / 200 \mathrm{~K} / 20$ MEG．OHM． E15．P．\＆P， 25 p ．
FTC－401 TRANSISTOR TESTER Full capabilities for mea－
suring $\mathbf{A}, \mathbf{B}$ and Suring A，B and ICO． adaptable for checking diodes．Supplied com－ plete with instructions， battery and leads．
$\mathbf{8 6 - 9 7 1}$ ．P．\＆P． 15

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Can be panel or bench nounted．Basic meter messures 1 volt D．C．but can be used to measure a wide range of $A C$ and DC volt， Gpecification：Accuracy： 10.0 piug in cards． Resolution： $1 \mathrm{~m} V$ ．Number of digits： 3 plus ourth overrange digit．Overrange： $100 \%$ （up to 1．999）Input impedance： 1000 Meg ohm．Measuring cycle： 1 per second．Adjust－ ment：Automatic zeroing，full scale adjust ment against an internal reference voltage Overloarl；to 100v．D．C．Input：Fully floating （3 poles）．Input power： $110-230 \mathrm{v}$ ．A．C． $\pm 83 / 16 \mathrm{in}$ ． AVAILABLE BRAND NEW AND FULLY
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Also see next two pages

# SEVI-EONDLOTORS/VALVES 

ALL DEVICES ERAND NEW AND FULLY GUARANTEED

| Transistor |  | - | 97 | 9N |  | 8C115 |  |  |  | NKT219 | 80 | Intetrated | FJH101 ${ }^{5} \mathrm{p}$ |  | VALYES |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 2N | 270 | 2N5 |  | ${ }_{8 C 116}^{\text {BC11 }}$ | $15 p$ | BFW91 |  | NK | $27 p$ |  | $\begin{array}{ll} \text { FJHI11 } & 70 p \\ \text { FJH121 } & 850 \end{array}$ |  |  |  |  |  |  |  |
| 2 C 302 | 180 | 2N | 180 | 28103 | ${ }_{20}$ | ${ }_{\text {BC118 }}$ | 15 p | BFX13 | 28p | NKT225 |  | CA3000 180p | FJH131 ${ }^{\text {20] }}$ | ${ }^{\text {S }}$ |  |  |  | 80 p |  |  |
| $2 \mathrm{G80}$ | 20 | 2N34 | 97 | 28104 | 250 | BC119 | 35 p | BFX29 | 25 p | NKT229 |  | CA3005 117 D | FJH141 20 | 3N7442 87p | OR | 10 | 25 | 650 | EMM 81 |  |
| $2 \mathrm{G306}$ |  | 2N356 | 170 | 28301 | 47 p | BC121 |  | BFX 30 | 255 | NKT237 | $8{ }^{85}$ | CA3007 288p | FJH151 \% ${ }^{\text {F }}$ | 8N7446 125p | IL4 | 20 p | 30C15 | 600 | EMM84 |  |
| 2 G 308 |  | 2N356 | 15p | 28302 | 470 | BC12 |  | BFX37 |  | NKT238 | 25p | CA3011 | FJH161 70D | 1100 | IR5 | P | 30 C 1 | 85 p | EM8 |  |
| 88 |  | 2N356 |  | 28803 | 57 | BCl2 | 15p | BFX44 | 87 p | KKT240 | 270 | CA3012 | FJH171 25p | SN7448 100 | IS5 | 5 p | 30C1 | 750 | EM87 |  |
| *37 | 16 | 2N35 | 25p | 28304 | 728 | BC1 | 259 | BFX 68 | 20 | NKT241 |  | Ca3013 10 | FJH181 25 p |  | TT4 | 851 | 30F5 | 85 p | WY51 |  |
| 2G374 | 20 | 2N3569 | 201 | 2 S 501 | 32 | $\mathrm{BCl}^{\text {B }}$ | 150 | BFX84 | 25 p | NKT242 | 20p | CA3014 184 | FJH221 25p | 8N7451 830 | 104 | 27 D | 30 FL 1 | 700 | EY86 |  |
| 2G881 | 29 | 2N3570 | 125 | 28502 | 35 p | BC13 | 15 p | BFX85 | 88 p | NKT |  | CA3018 84p | FJH281 250 | 8N7468 88p | IU5 | 60 p | 30FL12 | 98p | EY8 |  |
| 2N388 | 490 | 2N3572 | 978 | ${ }^{28503}$ | 27 | ${ }^{\text {BCl3 }}$ | 159 | BFX86 | 250 | NKT244 | 17 p | 8 A | FJH241 25 | SN744 | 2 D 21 | 85 p | $30 \mathrm{FL14}$ | 750 | EZ40 |  |
| 2N404 | 200 | 2N3605 | 27 | 3N88 | 40 p | BC137 | 150 | BFX87 |  | NKT240 |  |  | FJH251 259 | 857460 88p | $3 \mathrm{Q4}$ | 0 | 30L15 | 85 p | EZ41 |  |
| 2N69 | 15 p | 2N360 | 2 | ${ }^{3 N 128}$ | 70 p | ${ }^{\text {BCl }} 38$ | 87 p | BFX88 | 200 | NKT261 |  | CA3019 84p | FJJ101 ${ }^{\text {F0p }}$ | SN7472 85 | ${ }^{854} 4$. | , | $30 \mathrm{L17}$ | 80 | Ez80 |  |
| 2N69 | 15 p | 2Na60 |  | 3N140 | 77 | ${ }^{\text {BC140 }}$ | 35 | BFX89 |  | NKT262 | 200 | CA3020 1869 | FJJ111 800 | SN7473 48 | 3 V 4 | 45 | ${ }^{301212}$ |  | EZ81 |  |
| 2N69 |  | 2N363 | 18p | ${ }_{3}^{3 N 141}$ | 72p | ${ }^{\mathrm{BCl}} 41$ | 85 | BFX93A | 700 | NKT264 | 200 |  | FJJ121 | gN744 | 5 R 4 | 60 | 30719 | $80 p$ | GZ33 |  |
| 2N699 | 2. | 2N8638 |  | 3N142 |  | BC147 | 150 | BFY11 | 48p | NKT271 |  | 15 | FJJ181 | SN7475 47 D | 5 U 4 | 88p | 30PL1 | 70 p | GZ34 |  |
| 2N706 | 9 | 2N3641 | 18p | 3N143 | 67 | BC14 | 11. | BFY18 | 250 | NKT262 | 203 | CA3021 158 | FJJ141 185p | SN7476 ${ }^{\text {SN }}$ | 5 Y 4 | 48 | $30 \mathrm{PL1} 3$ | 98 | ${ }^{\mathrm{K} T 86}$ | 52.09 |
| 2N70 | 11 | 2N364 | 18 | 3N152 |  | BC149 | 150 | BFY19 |  | NKT274 | 20p | $\text { CA3022 } 180$ | FJJ 18180 | 9N7483 878 | 5 Y 3 | 88 | 30PL14 |  | KT88 | 3200 |
| 2N708 | 15p | 2N3643 | 20 p | 40050 |  | ${ }_{\text {BC15 }}$ | 770 | BFY21 | ${ }^{48}$ | NKT275 | 80 | $\text { CAB023 } 126$ | $\text { FJJ191 } 6$ |  | $5 \mathrm{Z4}$ | 40 p | 35L6 | 60 p | MU14 |  |
| 2N709 |  | 2N3644 |  | 40250 40251 |  | ${ }_{\text {BC154 }}$ | 5 | BFY24 | 5 | NKT278 | $25 p$ 870 | CA3026 10 | FJJ2611 12 | 6NN7490 100 | ${ }^{6 / 30}$ | 750 | 35 |  | PABC8 |  |
| 2N718A |  | 2N86 | 1 | 40309 |  | BC167 | 150 | BFY30 | 40 | NKT401 | 87 | B | FJL101 125p | SN7493 879 | 6AG7 |  | 3575 | 40 |  |  |
| 2N726 |  | 2N369 | 18 p | 40310 | 4 | BCl. 68 | 15p | BFY41 | 0 | NKT4 | 90p | 105p | FJY101 | 8N7495 87p | 6AK\% | 80p | 60B5 | 450 | PC97 |  |
| 727 | 25 p | 2N36 | 16 | 40311 | 859 | ${ }^{\text {BCl6 }}$ | 150 | BFY43 | 62 | MKT4 | 75p | CA3029 87p | IClo 250 p | gN7496 878 | 6AK6 | 570 | 50C5 | 40 p |  |  |
| 2N914 | 17 | 2N3694 | 18p | 40812 | 47 | ${ }^{\text {BCl }}$ | 85 p | BFY50 | 20 | NKT4 | ${ }^{\text {82p }}$ |  | IC12 250p | SN74107 489 | 6AL | 20 p |  | 500 | PCO84 |  |
| ${ }^{2 N 12}$ | 17 p | 2N3702 | 100 | 40314 | 37 | ${ }^{\mathrm{BCL}} 16$ | 15 | BFY51 | 20 | NKT40 | 75 p | 18 | L900 40p |  | 6AM6 | 8 p | 851 | 45 p |  |  |
| 2N918 |  | 2N8703 | 10 p | 40315 | 87 p | BC168 | 14 | BFY52 | 170 | NKT 406 | de | 187 | L914 409 | 14 | 6A | 35 p | 807 | 50 p | PCO88 |  |
| 2 S 929 | 29p | 2N3704 | 12 p | 40316 | 470 | ${ }_{8 C 168 C}^{\text {BCl }}$ | 15 | BFY53 | 17 p | NKT45 | ${ }^{69}$ | CA3035 128p |  |  | ${ }^{6486}$ | 370 | 1625 | 80 |  |  |
| 2N930 |  | 2N3705 | ${ }_{10 p}^{10 p}$ | 40317 40319 | 87 p | ${ }_{\text {BC169 }}{ }^{\text {BC16 }}$ | 140 | BFY56A | 42 | NKT4 | 82 | $372 p$ | M |  | 6AT6 | Op | ${ }^{5768}$ | 0 | PCC189 |  |
| 1090 |  | 2N370 | $12 p$ |  | 479 | BC170 | 12 p | BFY77 | 57 p | NKT713 | 20p | A3041 109 p | MC788P 82p |  | 6 6A | 80 | AZ81 | 500 |  |  |
| 2N1091 | 2 | 2®3708 | 8 p | 40323 | 82p | BCl7I | 159 | BFY90 |  | NKT717 | 48p | CA3042 109p | M0790P |  | 6 B | 250 | CY31 | 85 p | PCF84 |  |
| 1181 | 8 | 2N8709 | 10 p |  | 778 | ${ }^{\text {BC172 }}$ |  | B8x19 | 17 P | NET734 |  | CAS043 1879 | MC792P |  | 68 | 80 p | Daf91 | 250 |  |  |
| 2N130 |  | 2N3710 | 10 |  |  | ${ }_{8 \mathrm{BC17}}^{\mathrm{BC17}}$ |  |  | 17 p | N | ${ }^{5}$ | $\text { CA } 3044=180 \%$ |  |  | 6 BE 6 | 45 | DAF96 |  | PCF800 |  |
| 2N130 | $\begin{aligned} & \overline{17} \\ & 17 \end{aligned}$ | $\begin{aligned} & 2 \mathrm{~N} 3711 \\ & 2 \mathrm{~N} 3718 \end{aligned}$ |  |  |  | BC177 BC178 | p | ${ }_{\text {BSX } 21}$ | 20p | NKT7 | 809 | $\begin{array}{cc}\text { CA3046 } \\ \text { CA3046 } & \text { 129p } \\ \text { 81p }\end{array}$ | 家 | SN74165 ${ }^{\text {820p }}$ | ${ }_{68 \mathrm{Br}}^{6}$ | $\begin{aligned} & 45 p \\ & 400 \end{aligned}$ | ${ }_{\text {DF91 }}$ | $89$ |  |  |
| 2N1304 | ¢ | 2N3714 |  | 40347 | 57 p | BC179 | 280 | BSX27 | 47 p | OC16 | 60 p | CA3047 187p | MC1304P |  | 6 BR | 85 p | DK91 | 85 | PCF806 |  |
| 2N1305 |  | 2N3715 | 22 | 40348 | 68 p | BC182 | 12p | RSX28 | 82. | OC19 | 879 | CA3048 2049 |  |  | 6 BE | 650 | DK92 | 50 D | PCF806 |  |
| 130 |  | 2N3716 |  | 40 | 428 | BC182L | 100 | B8X60 | 82 | OC20 | 750 | $\begin{array}{ll} \text { CA3049 180p } \\ \text { OA30K0 } \\ 1850 \end{array}$ |  |  | $6 \mathrm{6B}$ | ${ }^{850}$ | DK96 | 48 | PCF808 |  |
| 13 |  | 2N3773 |  | 40361 | 47 D | ${ }^{\text {BC183 }}$ | 10 | B8X61 |  | $0022$ $0 \mathrm{Cl} 3$ |  | OA3050 1859 |  |  | 6 B | 0 p | DL92 | 85 p | PCLIS2 |  |
| 2N1308 |  | 2N3791 | 27 | 40362 |  |  | $\begin{aligned} & 10 p \\ & 160 \end{aligned}$ | $\begin{array}{\|c} \text { B8X76 } \\ \text { Bsx } 77 \end{array}$ | $\begin{aligned} & 150 \\ & 800 \end{aligned}$ | $\begin{aligned} & \mathrm{OC} 28 \\ & \mathrm{OC} 24 \end{aligned}$ |  | $\begin{array}{ll} \text { CA3051 } \\ \text { CA3052 } & 18 \end{array}$ |  |  | ${ }^{68}$ | 859 | DL94 | 459 | PCL83 |  |
| 2N1309 |  | 2N3819 |  | 4037 | 889 | $\begin{aligned} & \mathrm{BCl} \\ & \mathrm{BCl} \end{aligned}$ | $\begin{aligned} & 15 p \\ & 120 \end{aligned}$ | ${ }_{\text {BSX }}{ }^{\text {BS } 78}$ | $\mathbf{8 0 p}$ | $0 \mathrm{OC24}$ | 60 | $\begin{array}{ll} 052 & 1659 \\ 053 & 489 \end{array}$ |  |  |  | ${ }^{88} 15$ | DL96 |  | PCL84 |  |
| 16 | 81 | 2N3823 |  | 2407 | 4 | BC186 | 25 p | BSY24 | 15p | $0 \mathrm{C26}$ | 85 | 109p |  | 42 | GGL | 50 p | DY86 | 88 | PCL88 |  |
| 2F1681 | 85 | 2N3854 | 87 p | 40408 |  | BC187 |  | BSY25 | 159 | OC28 | 80 p | A3055 240 p |  |  | 6CP | 68 p | DY87 | 85 p | PFL20 |  |
| 2N1637 | 80 | ${ }^{2} \mathrm{~N} 3854$ | ${ }^{278}$ | 40409 |  | ${ }^{\text {BC2 }}$ B13L | 12p | $\left\lvert\, \begin{aligned} & \text { BSY26 } \\ & \text { BSY27 } \end{aligned}\right.$ | 17 p | ${ }_{\text {OC3 }}$ |  | CA3059 1659 |  | TAA243 160p | 6 FI | 82. | E8800 | 650 | Lus6 |  |
| 2 N 1637 |  | 2N3855 | 875 | 40 |  | $\begin{aligned} & \mathrm{BC} \\ & \mathrm{BC} \end{aligned}$ |  | B8Y28 | 178 |  |  | FCH101 850 |  | TAA263 76 |  | P |  |  | ${ }^{\text {PL81 }}$ |  |
| 2 N 1639 | 27 | 2N3 | 30 | 40467 |  | BCY1 |  | BSY29 | 170 |  | 2 | FCH111 105p | 000 P | TAA300 175p | 6 |  |  |  |  |  |
| 2N1701 | 110 | 2N38 | 85 y | 40488 | 35 p | BCY30 | , | BSY32 | 250 | OC4 | 95 | FCH121 105p | 11 | TAA310 125p | 6 F15 |  |  |  | PL84 |  |
| 1711 | 24 p | 2N3858 | 250 | 40628 | 72p | BCY31 |  | B8Y36 | 250 | O | 1 | FCH131 500 | PA222 487 D | TAA320 78 p | 6 Fr | 45p | EBC41 | 65 | PL500 |  |
| 11888 | 829 |  | 80 p |  |  | BCY32 | 0 p | BSY37 | 20, | OCA | 12p |  | PA230 1000 |  | T | 0 p | EbC81 |  | PL504 |  |
| 2N1893 | 87 | 2N38 | 270 | 40803 |  | BCY33 |  | B6Y38 | 20 | OC4 | 150 | P | PA234 1000 | $\begin{aligned} & \text { TAA4 } 85147 \mathrm{p} \\ & \text { TAA5 } \\ & \hline 1821 \end{aligned}$ | ${ }^{6} \mathrm{H}$ | P | EB |  | PY32 |  |
| 47 | 78 | 2N3859 |  | ${ }^{\text {AC107 }}$ | 800 | BCY34 |  | BSY39 |  | ${ }_{0}^{0070}$ | 150 |  | PA237 PA246 2050 |  | 6J4 | D | EB | 40 p | PY33 |  |
| 160 |  | $\begin{aligned} & \text { 2N } 3860 \\ & \text { NN } 8866 \end{aligned}$ |  |  |  | BCY38 |  | $\mathrm{BSY}^{\text {B8Y }} 51$ |  | ${ }_{0}^{0071}$ | 120 | FCH171 1051 | $\begin{array}{ll}\text { PA246 } & 2450 \\ \text { P4424 } & 8350\end{array}$ |  |  | 0 D | EBF |  | PY80 |  |
| T21.93 |  | 2N3 | 40 | ${ }^{\text {ACl28 }}$ |  | BCY40 | 60 p | BSY52 | 32 p | OC78 |  | FCH191 1050 | PA264 4470 | TAA811 445 |  |  |  |  |  |  |
| 2N2194 | 27p | 2N3877 |  | AC151 |  | BCY41 | , | B8Y5 | 870 | 00 | 80 D | FCH201 180p | PA265 497 P | TAB101 97 p | ${ }^{3}$ | 45p | EC88 |  | PY88 |  |
| 2194 |  | 2N39 | do | ${ }^{\text {ACl52 }}$ |  | BCY42 | 15 p | BSY54 | 0 | OC75 |  | FCE211 180p | SN7400 ${ }^{28 p}$ | $\begin{aligned} & \text { TAD100 150p } \\ & \text { TAD110 } 1978 \end{aligned}$ | K | 85 p | ECC40 | 6 | PY88 |  |
| 17 | 87 | 2N3900 | 40 | ${ }^{\text {ACl54 }}$ | 220 | $\mathrm{BCY}^{\text {B }} 3$ | $15 p$ | B8Y56 | 90 p | ${ }^{0} \mathrm{C}$ | 80 | FCH221 180p | 8N7401 | TAD110 197p | 6L6GT | 50 | ECC84 |  | PY800 |  |
| 2218 |  | ${ }^{2 N 3901}$ | 97 | AC176 | 29 | $\mathrm{BCY}^{\text {BCY }}$ | ${ }^{89}$ | BSY79 | 459 | 0 C 7 | 80 | 231.150p | 8N7402 818 | $\begin{aligned} & \text { BL408A } 187 \mathrm{p} \\ & \text { SL7020 } 147 \mathrm{p} \end{aligned}$ | LD20 |  | ECC |  |  |  |
| 22219 | 20 | 2N3908 |  | AC187 |  | ${ }^{\text {BCY }} 58$ |  | B8Y90 | 578 | ${ }^{0} \mathrm{Cl78}$ | 800 | 101160 p | 8N7403 ${ }^{\text {SN7 }}$ |  | $6 \mathrm{Q7}$ | 40 | ECC |  |  |  |
| 2 N 222 | 8 | 2N3 | 250 |  |  | $\begin{aligned} & \text { BCY } \\ & \text { BCY } 68 \end{aligned}$ | 9 | C4 | 12 p | 0081 |  | FCJ121 275 | 8N7405 |  | 68A |  | BCr80 |  | U26 |  |
| ${ }_{2} \mathbf{2 N} 2221$ |  | 2 N |  | ${ }^{\text {A }}$ |  | BCY | 15p | C450 | 150 | -C82 | 2 | FCO131 27 | SN7 | UA703C 187p | $\begin{aligned} & 68 \\ & 68 \\ & 7 \end{aligned}$ | 5 | ECFP88 |  | U50 |  |
| 2N2222A |  | 2N4 | 15 p | ACY19 | 24 p | BCY71 | 20 p | GET102 |  | OC82D |  | FCJ141 520p | 8N7408 28p | UA709C 125p | SK7 | 85 | ECH21 | 670 | 1 |  |
| 2 N 2297 | 30 | 2N40 | 10p | ACY20 | 200 | BCY72 | 15 p | GET113 | 20 p | $0 \mathrm{C83}$ |  | FCJ201 100p | 8N7409 88 s | UA710C 125p | 6857 | \% 5 p | ECE35 |  | U281 |  |
| 23 | 15p | 2N40 |  | ACY2 | 20 | BCY | 80p | GET114 | 15 | OC84 | ${ }^{250}$ | FCJ211 875 | 8N7410 | UA716 187p | 68N7 |  | ECH42 | 70 | U2 |  |
| 2N2369 |  |  |  |  |  |  |  | GET11 | 209 | ${ }_{0}$ |  | FCCL101 |  | UA7230 162p | 68 |  | RCB81 RCH83 |  | U8 |  |
| 2N2410 | 42 | 2N4244 | 47p | ACY39 | 470 | BCZ11 | 40 p | GET873 | 12p | OC170 | 2 | FCY101 10 | 77420 88p | VA741C 87p | - | 250 |  |  | UABC80 |  |
|  | 870 | 2N424 | 15p | ACY40 | 149 | BD112 |  | GET880 |  | 000171 | 80 |  |  |  | V6 | 82 | ECL82 |  | UAF42 |  |
| 2N2484 | 38. | 2N4249 | 15 p | ${ }^{\text {ACY }} 41$ | 25p | BD116 | 1120 | GET887 | , | OC200 | - | grcilitme |  |  | 6x | 80 p | ECL83 | 0 | UBC41 |  |
| 2N2539 | 220 |  |  | ACY44 | 70 | BD121 |  | GET89 | 22 D | $\mathrm{O}_{\mathrm{C} 222}$ |  | PLABTIO | ${ }_{200} 00$ PIV |  | X |  | EF3 | 40 | UBC81 |  |
| 2 N 2613 |  | 2N42 | 42 y | AD1 | 470 | BD124 | 750 | GET896 | 22p | OC203 | 400 | CAP | 400 PrV | 4A 80p | 100 | 7 | EF37 |  | UBF80 |  |
| 2 N2614 | 80 p | 2M428 | 170 | AD150 | 2 p | BD131 | 759 | GET89 | 28p | OC204 | 400 | 50 PIV 2A | ${ }^{60} 100$ PIV |  | 0F1 | 900 | EF40 |  | UCC84 |  |
| 281 | 470 | 2N4285 | 17 p | A | P | RD132 |  | GET88 |  | $00^{0} 2$ | 750 | 100 PIV 2 A . | 200 PIV | V6A 85p | 10P1 | 10 | $\mathrm{MF}^{\text {E1 }}$ | 65 | UCC85 |  |
| 2 N 2711 |  | 2N428 | 179 |  |  |  |  |  |  |  | ${ }^{90 \mathrm{p}}$ | 200 PIV 2 A | P) 400 PIV | V 6 A $\quad 810$ | $10 \mathrm{P14}$ | \$1-10 | ${ }_{\text {EF42 }}$ |  | UCFP\% |  |
| ${ }_{2712} 2713$ | 270 | 2N4288 | 150 | AF109 | 25 | BDY | 105p | MAT101 | Op | ${ }_{0}^{0} \mathbf{0} 2071$ |  | 200 Plv | , |  | 12 AT 6 | ${ }^{30} \mathrm{p}$ | EF80 | 25 | UCH21 |  |
| 2N2714 |  | 2N4289 | 16 | AF'15 | 5 | BDY62 | 1000 | Mat121 | 30 p | ORP12 | 5 |  | 8s |  | 12AU7 | 0p | EF86 | 30p | UCH81 |  |
| 2N2904 | 20 | 2N4290 | 15 y | AF116 | 2.50 | BF115 | 25 p | MJ400 | 1079 | ORP60 | 40 |  | Hex |  | $12 \mathrm{AX7}$ | 80 p | EF89 |  | UCL82 |  |
| ${ }^{2} \times 2904$ | 250 | 2N4291 | 159 | AF117 | 200 | ${ }^{\text {BF1717 }}$ | 470 | MJ 420 | 1189 | ORP61 | 420 |  |  |  | $12 \mathrm{AV6}$ | 88 p | EF91 | $3{ }^{1}$ | UCL83 |  |
| 2N290 | 25p | 2N4292 | 150 | AF118 | 49 | ${ }^{\text {BFIL }} 5$ |  | M, 421 | 1189 | P346A | 28, |  | SERITMS | Eme granims | 12 B | 85 | EF92 | 400 | UF41 |  |
| ${ }_{2 N}^{2 N 29006}$ | ${ }_{80}^{80}$ | 2N 422 | 179 |  |  | ${ }_{\text {BF15 }}{ }^{\text {BF15 }}$ |  | MJ480 | 108p | ST140 | 15 p |  | 1 AITP 1.5 A | AMP 8 AIPP | ${ }_{12 \mathrm{BE}}$ |  | EFi83 | 80 | UF80 |  |
| ${ }_{2}^{2 N 2906}$ |  | 2N430 | 479 | AF124 | 19p | ${ }_{\text {BFIL }}{ }^{\text {BF9 }}$ | ${ }^{670}$ | MS440 | 959 970 | TI884 | 880 | 4002 100PIV |  | 10p ${ }_{\text {19p }}$ |  |  | EF184 EH90 | 35 p | UF89 |  |
| 2N2907 | 230 | 2N496 | 182 | AF126 | 165 | BF168 | 85 p | MJ481 | 1250 | Tr843 | 409 | 4003 200PIV | 10 p | 28 | 20 Dl | 45p | EL34 |  | UL41 |  |
| 2N292 | 150 | 2N5027 | 52p | AF127 | 169 | BF167 | 25 p | M.J490 | 100 p | TIS4 | 18 p | 004 400PIV | 10 p 18 | 20 250 | 20 F 2 | 76 p | EL33 | 31.20 | UL84 |  |
| $2 \mathrm{2N} 2924$ | 150 | 2N502 | 37 D | AF1 | 20 | BF | 3 p | MJ491 | 18 | TI | 18 | 005 600PIV | 18p | 10 p 80p | 20 L 1 | 41.10 | EL41 | 56 | UY41 |  |
| 2NT2925 | 15 | 2N5029 | 470 420 |  |  |  |  | MJE340 |  |  |  |  |  |  | $20 \mathrm{P1}$ | 50 p | Li42 | 6 | UY85 |  |
| 2N2926 | 18p | 2N5030 | 42p | AF179 AF180 | 5 | ${ }_{\text {BF178 }}$ | 50 | MJJ370 | 957 870 | TTR48 | 189 $18 p$ | 1000 P | e8s 15\% 100 + | 00p $20 \%{ }^{\text {80p }}$ | $20 \mathrm{P3}$ | 80 | EL8 | 659 | VR1 |  |
| 2N2926Y | 18 | 2N517 |  | AF181 |  | BF179 |  | MJE620 | 875 | TIS49 | 18 p |  |  |  |  |  |  |  |  |  |
| 2N3011 | 24 | 2N5175 | S20 | AF186 | 89 | BF180 | 850 | Mse521 | 870 | TIS80 | 17p |  | O10 | FR |  |  |  |  |  |  |
| 2 N 3014 |  | 2N5176 | 45 D | AF239 | 800 | BF181 | 85 p | MPF102 | 420 | Ti851 | 189 |  | UD Moul | 17 |  |  |  |  |  |  |
| 2 N 305 | 20 p | 2N5232 |  | AF279 | 4 | BF182 | 80 p | MPF108 | 85p | TIS | 18p |  |  |  | D1 |  | E |  |  |  |
| 2N305 | 49 | 2N5245 | 45 | AF280 | 470 | BF184 | 200 | MPF104 | 87 D | TIS63 | 89 | 00PIV |  | 60p 81.48 | 1N34A | 10 p | BA154 |  | 73 |  |
| 2N3055 | 28 | 2 N 52 | 429 | AFZ11 | 88 p | BF185 | 209 | MPF105 | 8 | XB112 | 18 p |  | 5 p | (10p 81.48 | 1N914 | 7 | BAX13 | 12 D | OAS |  |
| 2N3183 | 25 | 2N5249 | 67 | A8Y26 | 850 | BF194 | 170 | MPS36 | 82p | XCl41 | ${ }^{85}$ | OPIV |  | 629 ${ }^{68}$ | 1 N 916 | 7 D | BAX16 | 7 P | OA6 |  |
| 2 N 3134 | 80 p | 2N52 | 825p | A8Y27 | 81 | BF195 | 159 | NKT124 | 480 | ZTX107 | 110 | 00PIV | 76 | 7870 | AA119 | 7 D | BAY31 | 7 p | OA10 |  |
| 2 N 3135 | 25 | 2N530 | 87 D | ASY28 | p | BF196 | 15p | NKT126 | 870 | ZTX108 | $11 p$ | 0 PIV |  |  | AA129 | 10p | BAY ${ }^{\text {B }} 100$ | 17 p | OA9 |  |
| 2N3136 | 25 D | 2N58 | 20 | A8Y29 | 4 | BF197 | 150 | NKT126 | 870 | ZTX109 | 118 | ${ }_{50}+$ les | 15\% 100+ les | 2088 $20 \%$ | AAZ13 | 10 p | BY100 | 15p | OA47 |  |
| 2N3390 | 25 | 2N5307 | 87 p | ASY50 | 250 | BF198 | 48p | NKT128 | 870 | ZTX300 | $11 p$ | $50+1$ | 15\% 100+ les | S80\% | AAZ15 | 120 | BY103 | 28p | OA70 |  |
| 2N3391 | 20 p | 2N530 | 870 | AEY51 | 88 p | BF200 | ${ }^{35}$ | NKT135 | 270 | ZTX301 | 170 |  | 表 |  | BA100 | 15 p | BY122 | ${ }^{878}$ | 0473 |  |
| 2 N 8391 | 1 | 2N5308 | ${ }^{8 \%}$ | A8Y54 | ${ }^{2}$ | $\mathrm{BF}^{\text {Br224 }}$ | ${ }^{20 p}$ | NKT137 |  | ZTx 802 |  |  |  |  | BA102 |  | ${ }_{\text {BYY12 }}$ |  | OA79 |  |
| 2 N 3392 | 17p | 2N5310 | 48 | Agye 7 | 459 | ${ }^{\text {Br }} 2225$ | 200 | NKT210 | 80p | ZTX303 | ${ }_{98}^{18 p}$ |  | $24-100$ | 39-9-100V | BA110 | 889 | BY128 | 170 | OA81 |  |
| 2 N 3398 | 15p | 2N5354 | 87 P | A8Y86 | 88p | BF287 | 298 | NKT211 | 80p | ZTX304 | ${ }^{258}$ |  |  | 25 y each | BA111 | 870 | BY127 | 17 p | OA85 |  |
| 2 N 3384 | 15p | 2N5855 | 27 | AEZ21 | 870 | BF238 | 280 | NKT212 |  | ZTX 500 | 189 | $25+$ less | $15 \% 100+$ less | 20\% | BA112 | 70 p | BY164 | 87p | OA90 |  |
| 2 N 8402 | 82p | 2N5356 | 82 | AUY10 |  | BF244 | 82p | NKT213 | 80p | ZTX 501 | 18p | \% less | 10\% $100+1$ ess | 20\% | BA115 | 7 p | BY210 |  | DA91 |  |
| \$8408 |  | 2 |  | BC107 | 100 | BFW61 | 470 | NKT214 | 220 | ZTX 502 | 80 p | TaA | ,00 Vatis:- | 2 $+10 \%$ | BA141 | 889 | BYZ11 |  | OA95 |  |
| §3404 | 88ip | 2N5866 | 820 | BC | 109 | BFW87 | $25 p$ | NKT215 | 229 | ZTX503 | 17 p | 25 | any | type. Post- | BA148 | ${ }^{88 p}$ | ${ }_{\text {BYZ12 }}$ | \% | OA200 |  |
|  | 22p | 2N5367 |  |  | 100 | BFW88 | 289 | NKKT216 |  | ZTX504 | 40p | A.E. YOE | ULL Liszs. |  | BA144 | 120p | BYYZ18 | 80p | OA203 | 170 |

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Power Gain 9.0dB (1,000,000,000 times) after feedback.
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Input Impedance 250 Kohms nominal.
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With the addition of only a very few external resistors and capacitors the Super IC. 12 makes a complete high fidelity audio amplifier suitable for use with pick-up. F.M. tuner etc. Alternatively, for more elaborate systems, modules in the Project-60 range such as the Stereo 60 and A.F.U. may be added. The comprehensive manual supplied with each unit gives full circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include car radios, oscillators etc. The very low quiescent consumption makes the Super IC. 12 ideal for battery operation.


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Typical Project 60 applications

| System | The Units to use | together with | Cost of Units |
| :---: | :---: | :---: | :---: |
| Simple battery record player | 2.30 | Crystal P.U., 12 V battery volume control | £4.48 |
| Mains powered record player. | Z.30, PZ. 5 | Crystal or ceramic P.U. volume control etc. | ¢.9.45 |
| $20+20 \mathrm{~W}$. stereo amplifier for most needs | $\begin{aligned} & 2 \times 2.30 \text { s, Stereo } 60, \\ & \text { PZ.5 } \end{aligned}$ | Crystal, ceramic or mag. P.U.. F.M. Tuner. etc. | . $£ 23.90$ |
| $20+20 \mathrm{~W}$. stereo amplifjer with high. performance spkrs. | $\begin{aligned} & 2 \times 2.30 \text { s, Stereo } 60, \\ & \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U., F.M. Tuner. Tape Deck, etc. | £26.90 |
| $40+40 \mathrm{~W}$. R.M.S. de-luxe stereo amplifier | $2 \times 2.50$ s, Stereo 60 PZ.8, mains trsfrmr | As above | £34.88 |
| Indoor P.A. | Z.50, PZ.8, mains transformer | Mic., guitar, speakers, etc., controls | £19.43 |

[^5]
# from a simple amplifier to a complete stereo tuner amplifier with Project 60 modules 

## Z. 30 \& Z. 50 power amplifiers



The $Z .30$ and $Z .50$ are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at full output and all lower outputs. Whether you use $Z .30$ or $Z .50$ amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.
SPECIFICATIONS ( $Z .50$ units are inter-
changeable with $Z .30$ s in all applications).
Power Outputs
Z. 3015 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 30 hms using 30 volts.
2.5040 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts.
Frequency response: 30 to $300,000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$.
Distortion: $0.02 \%$ into 8 ohms.
Signal to noise ratio: better than 70 dB unweighted. Input sensitivity: 250 mV into 100 Kohms .
For speakers from 3 to 15 ohms impedance.
Size: $14 \times 80 \times 57 \mathrm{~mm}$.
2.30

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£4.48
2.50

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£5.48

## Power Supply Units

Designed special for use with the Project 60 system of your choice. Use PZ. 5 for normal Z. 30 assemblies and PZ. 6 where a stabilised supply is essental.
PZ.5 30 votts unstabilised $\mathbf{£ 4 . 9 8}$ PZ. 635 volts stabilised $£ 7.98$ PZ. 845 volts stabilised (less mains transformer) $\mathbf{£ 7 . 9 8}$ PZ.8 mains transformer $£ 5.98$

## The Sinclair Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guoranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.

## Project 60 Stereo F.M. Tuner

First in the
world to use the
phase lock loop principle

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now. Sinclair have appled the principle to an F.M. tuner with fantastically good results. Other original features include varicap dıode tuning. printed circuit coils. an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. Good reception is possible in difficult areas. and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high fidelity system.
SPECIFICATIONS-Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz . Capture ratio: 1.5 dB . Sensitivity: $2 \mu \mathrm{~V}$ for 30 dB quieting: $7 \mu \mathrm{~V}$ for full limiting. Squelch level : $20 \mu V$. A.F.C. range: $\pm 200 \mathrm{KHz}$. Signal to noise ratio: $>65 \mathrm{~dB}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}( \pm 1 \mathrm{~dB}$ ). Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation. Stereo decoder operating level: 2 NV . Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S.
Operating voltage: 25-30 VDC. Indicators: Mains on ; Stereo on ; tuning. Size: $93 \times 40 \times 207 \mathrm{~mm}$.
£25

## Stereo 60 Pre-amp/control unit <br> 

Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.
SPECIFICATIONS-Input sensitivities: Radio-up to 3 mV . Mag. p.u. 3 mV : correct to R.I.A.A curve $\pm 1 \mathrm{~dB}: 20$ to $25,000 \mathrm{~Hz}$. Ceramic p.u. -up to 3 mV : Aux - up to 3 mV . Output: 250 mV . Signal to noise ratio: better than 70 dB . Channel matching: within 1 dB . Tone controls: TREBLE +75 to -15 dB at 10 KHz : BASS +15 to -15 dB at 100 Hz . Front panel : brushed aluminium with black knobs and controls. Size: $66 \times 40 \times 207 \mathrm{~mm}$. Built tested and guaranteed.
£9.98

## A.F.U. High \& Low Pass Filter Unit



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid ( $12 \mathrm{~dB} /$ octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages - rumble (high pass) and scratch (low pass). Supply voltage -15 to 35 V . Current -3 mA . H.F. cut-off ( -3 dB ) variable from 28 KHz to 5 KHz . L. F. cut-off ( -3 dB ) variable from 25 Hz to 100 Hz . Distortion at $1 \mathrm{KHz}(35 \mathrm{~V}$. supply ( $0.02 \%$ at rated output. Size: $66 \times 40 \times 90 \mathrm{~mm}$.

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OA3 | 0.45 | 6AS5 | 0.45 | 6 F 13 | 0.45 | 10F1 | $0.75$ |  |  |  |  |  |  |  | $\begin{aligned} & \text { EP98 } \\ & \text { WFI83 } \end{aligned}$ | $\begin{aligned} & 0.65 \\ & 0.30 \end{aligned}$ | $\begin{aligned} & \text { GZ34 } \\ & \text { HABCB } \end{aligned}$ | $\begin{array}{r} 0.60 \\ 800.45 \end{array}$ | $\underset{\text { PTRY }}{\underset{\text { PL }}{2}}$ | $\begin{aligned} & 0.45 \\ & 0.45 \end{aligned}$ | U301 | $0.40$ |
| OB2 | 0.35 | 6AS7G | 0.85 | 6 F 14 | 0.76 | 10F9 | 0.60 | First | fity |  | IY | 1 | In | 0 |  | $\begin{aligned} & 0.80 \\ & 0.85 \end{aligned}$ |  | $\begin{gathered} 800.45 \\ 0.40 \end{gathered}$ | PL83 | $0.45$ | U403 <br> U404 | $0.60$ |
| OB3 | 0.70 | 6AT6 | 0.35 | 6 F 15 | 0.65 | 10 F 18 | 0.50 |  |  |  |  |  |  |  | EF800 | 0.85 1.00 | HK96 | 0.40 2.05 | ${ }_{\text {PLS402 }}$ | 0.40 0.85 | $\begin{aligned} & \mathbf{U 4 0 4} \\ & \mathbf{T} 801 \end{aligned}$ | $\begin{aligned} & 0.60 \\ & 0.80 \end{aligned}$ |
| 0 OS | 0.38 | 6AU6 | 0.25 | 6 F 18 | 0.50 | 10L1 | 0.50 |  |  |  |  |  |  |  | EK90 | 0.30 | KT66 KT88 | 2.05 2.00 | $\begin{aligned} & \text { PL302 } \\ & \text { PL504 } \end{aligned}$ | $\begin{aligned} & 0.85 \\ & 0.80 \end{aligned}$ | U801 <br> UABC8 | $0.80$ |
| OD3 | 0.85 | 6AV6 | 0.80 | 6 F 23 | 0.85 | 10 LDIl | 0.65 |  |  |  |  |  |  |  | EL33 | 1.35 1.25 | K788 N78 | 2.00 1.50 | $\begin{aligned} & \text { PL504 } \\ & \text { PL508 } \end{aligned}$ | $\begin{aligned} & 0.80 \\ & 0.90 \end{aligned}$ | UABC80 |  |
| 1B3GT | 0.40 | 6AW8A | A 0.60 | GF24 | 0.75 | 10P13 | 0.60 |  |  |  |  |  |  |  | EL34 | 1.25 | N78 | 1.60 00.40 | PL508 | 0.90 1.10 | UAF41 | 0.40 |
| ${ }_{1 L 4}^{1 \mathrm{CP} 31}$ | 7.00 0.20 | 6BA6 | 0.25 0.80 | 6F25 | 1.00 | 10P14 | 1.10 |  |  |  |  |  | N0 |  | EL36 | 0.50 | PC86 | 0.60 | PL801 | 0.80 | UAF42 | 0.55 |
| 1R4 | 0.45 | 6 BF 5 | 1.00 | 6 F 28 | 0.65 | 12AC6 | 0.50 |  |  |  |  |  |  |  | EL37 | 1.80 | PC88 | 0.60 | PL802 | 0.95 | UBC41 | 0.50 |
| 1R5 | 0.40 | 6BF6 | 0.55 | 6 J 4 | 0.50 | 12AD6 | 0.55 |  |  |  |  |  |  |  | EL41 | 0.60 | PC97 | 0.50 | PM84 | 0.60 | UBC81 | 0.40 |
| 184 | 0.30 | 6BH6 | 0.75 | 6 J G9 | 0.80 | 12aE6 | 0.55 |  |  |  |  |  |  |  | ELA2 | 0.65 | PC900 | 0.48 | PY31 | 0.80 | UBF80 | 0.40 |
| 195 | 0.30 | 6 BJS | 0.50 | 6 J 6 | 0.20 | 12AL5 | 0.50 |  |  |  |  |  |  |  | HL81 | - | PCC84 | 0.40 | PY33 | 0.88 | UBF89 | 0.85 |
| 1T4 | 0.30 | 6BK7A | 0.60 | 6 J 7 | 0.45 | 12AQ5 | 0.50 |  |  |  |  |  |  |  | EL83 | 2 | PCC85 | 0.40 | PY80 | 0.40 | UBLI | 0.60 |
| 1U4 | 0.80 | 6BN5 | 0.48 | 6 K 6 GT | 0.60 | 12AT6 | 0.80 |  |  |  |  |  |  |  |  | - | PCC88 | 0.55 | PY81 | 0. | BL21 | 0.65 |
| 1U5 | 0.60 | 6BN6 | 0.45 | 6 K 7 | 0.35 | 12AT7 | 0.85 | 30 Cl 70.90 | 90 Al | 2.40 | DS87 | 0.8 |  |  | EL86 | 0.48 | PCC1 | 5 | 82 | 0.85 | C92 | 0.40 0.40 |
| 1V2 | 0.50 | 6BQ5 | 0.25 | 6K8G | 0.40 | 12AU6 | 0.85 | 30 Cl 80.80 | 90AV | 2.50 | DY802 | 0.85 | ECF | 0.35 | EL90 | 0.38 | PCC80 | 0.85 | PY88 | 0.38 0.40 | UCF80 |  |
| 1X2B | 0.50 | 6RR8 | 0.70 | 6 K 25 | 0.75 | 12AV7 | 0.80 | $30 \mathrm{F5}$ 0.85 | 90 CI | 0.65 | E88CC | 0.65 | EOF8 | 0.85 | EL95 | 0.85 | PCC80 | 0.80 | PY500 | 1.00 | UCH21 | 0.60 |
| 2D21 | 0.38 | 68S7 | 1.80 | ${ }_{616 G T}$ | 0.45 | 12AV6 | 0.40 | $30 \mathrm{FL1} 10.75$ | 90CV | 2.40 | E180F | 1.00 | ECF8 | 0.65 | EL821 | 0.60 | PCF80 | 0.80 | PY800 | 0.40 | UCH42 | 0.70 |
| 3A4 | 0.40 | 6BW6 | 0.85 | 6 L 7 | 0.40 | 12AV7 | 0.65 | 30 FL 121.20 | 807 | 0.50 | E810F | 2.90 | ECF8 | 1.65 | EL822 | 1.30 | ${ }^{-} \mathrm{CF} 82$ | 0.35 | PY801 | 0.50 | पCH81 | 0.40 |
| ${ }_{384}{ }^{38} 4$ | 8.00 0.35 | 6BW7 | 0.80 | 6L18 | 0.45 | 12AX 7 | 0.80 | 30 FLL 40.85 | 813 | 8.75 | EABC80 | 0.85 | ECH | 0.75 | ELL80 | 0.75 | PCF84 | 0.60 | PZ30 | 0.35 | UCL81 | 0.80 |
| 3V4 | 0.48 | 6BZ6 | 0.40 | 6N7G | 0.45 | 12AY7 | 0.75 | 30 LL 1 0.40 | B66A | 0.75 | EAF42 | 0.55 | ECH8 | 0.80 | EM34 | 1.00 | PCE86 | 0.60 | QqV02 |  | UCL82 | 0.35 |
| 5R4GY | 0.75 | 6C4 | 0.38 | 6 P 1 | 0.60 | 12BA6 | 0.60 | ${ }^{30 \mathrm{LI}} 00.85$ | 6642 | 0.70 | EBC33 | 50 | ECH8 | 0.45 | E | 0.80 | PCF87 | 0.80 |  | 2.25 | UCL83 | 0.60 |
| 5U4G | 0.35 | 6C5GT | 0.40 | 6P28 | 0.65 | 12BA7 | 0.40 | 30P12 0.80 | 60 | 1.60 | EBC41 | 0.55 | E | 0.45 | EM80 | 0.45 | PCE80 | 0.50 | QQvo3 | 10 | UF4I | 0.60 |
| 5V4G | 0.45 | 6 CB 6 | 0.36 | 607 | 0.49 | 12BE6 | 0.40 | $\begin{array}{ll}30 \mathrm{P} 19 & 0.85\end{array}$ | 6146 9360 | 1.60 | EBC81 | 0.80 | ECL8 | 0.45 | E | 0.60 | PCF8 | 0.50 |  | 1.25 | UF42 | 0.60 |
| 5Y3GT | 0.40 | 6CD6GA | 41.25 | 6SA7 | 0.40 | 12BH7 | 0.45 | $30 \mathrm{PL1} 0.75$ | 6939 | 1.25 | EBF80 | 0. | ECL8 | 0.50 | E | 0.35 | PCF | .80 | QQV03 | 20A | UF43 | 0.60 |
| 523 | 0.60 | 6CG7 | 0.55 | 6SG7 | 0.40 | 12BY7 | 0.60 | $30 \mathrm{PL13} 0.98$ | 7199 | 0.80 | EBF89 |  | ECL8 |  | EM85 | 1. |  | 0 |  | 5.25 | UF80 | 0.35 |
| 5Z4G | 0.40 | 6 CH 6 | 0.60 | 68K7 | 0.40 | 12 K 5 | 0.70 | 30PL14 0.90 | 7360 | 2.00 | EC53 | 0.50 | ECL8 |  |  |  |  |  |  | -40A | 5 | 0.40 |
| 6/30L2 | 0.80 | 6CL6 | 0.55 | 68L7GT | 0.85 | 12K7GT | '0.40 | 35 A 30.55 | 7586 | 1.85 | EC86 | 0.80 | ECL8 |  | EN | 0. |  |  |  | 5.50 | 9 | 0.40 |
| $6 \mathrm{AB4}$ | 0.85 | 6CW4 | 0.65 | 6SN7GT | 0.85 | 12Q7GT | 0.40 | $35 \mathrm{A5} 50.75$ | 7895 | 1.25 | EC88 | 0.80 | FCL8 |  | EYO1 |  |  |  |  | 3.00 |  | 0.85 |
| 6 AF 4 A | 0.55 | $6 \mathrm{CY5}$ | 0.45 | 6SQ7 | 0.40 | 12887 | 0.40 | 3585 0.65 | 9002 | 0.60 | EC90 | 0.88 | ECLI |  | EY81 | 0.55 |  | 0. |  | 8.20 |  | 0.40 |
| 6AG7 | 0.40 | 6 CY 7 | 0.70 | 6SR 7 | 0.40 | 20D1 | 0.50 | 35 C 50.50 | 9003 | 0.50 | EC91 | 0.60 |  | 1.85 | EY83 | 0.45 | PCL88 | 0.65 | TY2-12 |  | UM | 0.85 |
| 6AH6 | 0.50 | $6 \mathrm{D3}$ | 0.50 | $6 \mathrm{T8}$ | 0.35 | $20 \mathrm{L1}$ | 1.10 | 35 D 50.75 | AZ1 | 0.55 | EC92 | 0.35 | EF40 | 0.50 | EY86 | 0.40 | PCL85 | 0.40 |  |  | U88 | 0.60 |
| 6aj8 | 0.30 | 6DC6 | 0.80 | 6U4GT | 0.65 | 20 Pl | 0.50 | 35L6GT0.50 | AZ31 | 0.55 | EC93 | 0.55 | EF41 | 0.85 | EY87 | 0.40 | PCL86 | 0.40 | U18/20 | 0.75 | UY1N | 0.50 |
| 6 6K5 | 0.35 | 6DK6 | 0.50 | 6U8A | 0.40 | 20P4 | 1.10 | 35W4 0.35 | CBL1 | 0.90 | EC8010 | 2.25 | EF42 | 0.70 | EY88 | 0.48 | PCL88 | 0.45 | U26 | 0.80 0.80 | UY41 | 1.00 |
| 6AK6 | 0.60 | 6DQ6B | 0.70 | 6V6GT | 0.40 | 20P5 | 1.20 | $35 \mathrm{Z3} \quad 0.70$ | CBL31 | 1.00 | ECC40 | 0.65 | EF80 | 0.25 | EZ40 | 0.50 | PCL800 | 0.98 | U31 | 0.80 | UY82 | 0.48 |
| 6 AL 3 | 0.48 | 6DS4 | 0.75 | $6 \mathrm{X4}$ | 0.85 | 25C5 | 0.50 | 35Z4G 0.35 | CY31 | 0.85 | ECC81 | 0.35 | EF83 | 0.55 | EZ41 | 0.50 | PCL80 | 0.75 | U37 | 2.10 | UY85 | 0.60 0.40 |
| 6AL5 | 0.20 | 6EA8 | 0.60 | 6X5GT | 0.40 | $25 L 6 \mathrm{GT}$ | 0.50 | 3525GT 0.60 | DAF96 | 0.45 | ECC82 | 0.30 | EF85 | 0.35 | EZ80 | 0.27 | PD500 | 1.30 | U52 | 0.35 | W729 | 0.75 |
| 6AM5 | 0.35 | ${ }_{6 E 57}$ | 0.30 0.35 | 6X8 | 0.80 0.70 | 2524G | 0.30 | $\begin{array}{ll}\text { 60A5 } & 0.75\end{array}$ | DF96 | 0.45 | ECC83 | 0.30 | EF86 | 0.80 | EZ81 | 0.29 | PF86 | 0.65 | U76 | 0.35 | Z759 | 8.00 |
| 6AM6 | 0.33 0.38 | ${ }_{6}^{6 \mathrm{E}, 5} 7$ | 0.35 0.70 | 6Y64 | 0.70 0.65 | 2526GT | 0.65 | $\begin{array}{ll}50 \mathrm{B5} & 0.50 \\ 50 \mathrm{C} 5 & 0.50\end{array}$ | DK40 | 0.60 | ECC84 | 0.30 | EF89 | 0.88 | GY501 | 0.80 | PF818 | 0.85 | U78 | 0.35 | Z803U | 1.20 |
| 6AQ6 | 0.85 | 6F1. | 0.75 | 98W\% | 0.60 | 30AE3 | 0.40 | 50CD6G1.20 | DK92 | 0.55 0.45 | ECC88 | 0.40 0.40 | EF91 | 0.83 0.35 | GZ30 | 0.40 | PFL200 | 0.65 | U191 | 0.75 |  |  |
| 6AR5 | 0.45 | 6F5 | 0.75 | 10 C 2 | 0.50 | 30 Cl | 0.80 | 60L6GT0.55 | DM160 | 0.65 | ECC89 | 0.40 | EF92 | 0.35 0.35 | GZ31 | 0.35 0.48 | PL33 | 0.35 0.55 | U201 | 0.85 0.40 |  |  |
| 6AR6 | 0.55 | 6F6G | 0.35 | 10D1 | 0.55 | 30 Cl 5 | 0.80 | 85A2 0.50 | DY86 | 0.32 | ECC91 | 0.20 | EF97 | 0.65 | GZ33 | 0.70 | PL81 | 0.50 | U282 | 0.40 0.40 |  |  |

## 2N69 <br> TRANSISTORS

2N706
2N708
2N753
$2 N 753$
$2 N 929$
2N930
2N997
2N1131
2N1132
2N1184
2N1301
2N1302
2N1304
$2 N 1304$
$2 N 1305$
2N1306
2N1306
2N 1307
2N1307
2N1308

2N1309 \begin{tabular}{l|ll}
.17 \& ACl28 \& 0 <br>
ACl32 \& 0 <br>
\hline \& AC153 \&

 

0.70 \& A <br>
0.10 \& $A$ <br>
\hline
\end{tabular}

## S



## SYNCHROSCOPE TYPE C1-5



3-in tube fitted with telescopic verning hooit, giving bright display in full daylight.
Bandwidth $10 \mathrm{c} / \mathrm{s}-10 \mathrm{me} / \mathrm{s}$. Triggered sweep pre-set at
$1-2-5-10-30-100-300-1000-3000$ usec $1-2-5-10-30-100-300-1000-3000$ usec per stroke. Free-
running time base $20 \mathrm{o} / \mathrm{s}$ to $2000 \mathrm{kc} / \mathrm{s}$ with buitt-in crystal calibrator providing timing marks at . $05-.2-1-5$ -
Amptitude calibrator directly calibrated in volts.
input attenuator $1-10-100$.
PRICE $\mathbf{2 s 9 0}^{29}$. Packing and carriage 1.50

MINIATURE CERAMIC CAPACITORS 25V WOREING
$5 \% \quad$ tolerance $22-27-33-39-47-56-68-82-100-120-150-$ All $50 \%-20 \%$ tolerance $1500-2200-3300-4700-6800-10,000$ $-15,000 \mathrm{pF}-$ All at 83 np per twenty,
$80 \%-20 \%$ tolerance $0.022-0.033-0.047 \mu \mathrm{~F}$.
All at $28 n \mathrm{n}$ per twenty.
Minimum orders 20 of one size.

10\% TOL. METAL CASED PAPER CAPACITORS 250 V DC: $0.05-0 \cdot 10-0.25-0.5-1 \cdot 0 \mu \mathrm{~F}$ : 81.00 per twenty 500 V DC: $0.025-0.05-0.10-0.25-0.50 \mu \mathrm{~F}$ \&1.25 per twenty

Minimum Orders: 20 Capacitors

10-AMP SLLICON STUD MOUNTED RECTIFIERS $\begin{array}{lll}\text { BYX42- } 300 \text { piv } & 0.35 \\ \text { BYX } 42-600 \text { piv } & 0.40\end{array}$
$\begin{array}{lll}\text { BYX42- } \\ \text { BYX42- } 600 & \text { piv } & 0.40 \\ \text { BYX42- } 900 \text { piv } & 0.55\end{array}$ All these are available also


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Type U4312 U4313
with AC current ranges
Type U4812-Low sensi-
tivity instrument (667 tivity instrument ( 667
o.p.v.) $1 \%$ aceuracy on
nges on DC ( $300 \mu \mathrm{~A}$ to $\mathbf{~} \mathrm{OA}$ ) DC and $1.5 \%$ on AC. 9 current ranges on DC ( $300 \mu \mathrm{~A}$ to CA )
and 8 current ranges on AC ( $1 \cdot \mathrm{mmA}$ to 6 A$) .9$ voltage ranges and 8 current ranges on $\mathrm{AC}(1.5 \mathrm{~mA}$ to 6 A$)$. 9 voltage rang
0.3 to $900 \mathrm{~V} A \mathrm{C} / \mathrm{DC}$. Resistance 0.2 to $30 \mathrm{k} \Omega$. Mirror scale.
PRICE, complete with leads and steel carrying case $\mathbf{2 9 . 7 5}$ Type U4318-20.000 o.p.v. DC and 2,000 o.p. $\mathbf{8}$. on AC. 8 DC current ranges $60 \mu \mathrm{~A}$ to $1.5 \mathrm{~A}, 6 \mathrm{AC}$ current ranges $0-6$ to $1-5 A .9$ DC and AC voltage ranges 1.5-500V. Resiat-
Mirror scale.
PRICE, complete with test leads and steel carrying case case
$\mathbf{\$ 1 0 . 5 0}$

## * II: AG/DG MULMMETER AN <br> TRANSISTOR TESTER 8-DC vollage ranges $0.3-900 \mathrm{~V}$

$6-\mathrm{AC}$ voltage ranges $1.5-700 \mathrm{~V}$ 5-DC current ranges $0.06-600 \mathrm{~mA}$ $4-\mathrm{AC}$ current ranges $0.3-300 \mathrm{~mA}$ 4-resistance ranges $0.5-500 \mathrm{Kohms}$
Transistor cut-off current $60 \mu \mathrm{~A}$ max Transistor cut-oir current $60 \mu \mathrm{~A}$ max Sensitivity: $16700 \Omega / \mathrm{V}$ DC; $3300 \Omega / \mathrm{V}$ AC
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C. \& G. Eng. Crafts
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