## - A HI-FI A.M. RECEIVER  Hisw

## CONTENTS

a BEGINNER'S CONSTRUCTIONAL COURSE A MAINS/BATTERY PORTABLE

A TAPE ECONOMISER
A FREQUENCY COMPARATOR
TRANSISTORS IN PRACTICE SERVICING THE COSSOR RECORD PLAYER


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 19＇9 Consistung of Mains Mans－ 299 former．F W．Blduse Metal 299 Rertifier．well ventidated＝tan）
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 Consasting of F．W．Bridge
 Гrans．0－9－15 \％，but put and
Anmeter． 49 ．POst Ammeter． 499. Post 39.

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All for A．C．Mains $200-250$ v．，Socics． Guaranteed 12 months．


## Assembled 6 v ．

 or 12 v． 4 amps． Filted Ammeter and variable ，harke late elextole Almo seler－ t．01 plug lor 6 12ty 6 harging．Tin！ fed ileel rase wit li stoved blue hammer finush．Fused $5 /-$ use with mains and ant put leads．Carr． 16．Credit Terma： begnait 30－and－+ munthy pas nitut： 13
## R．S．C．MAINS TRANSFORMERS（nerit（1才）



 250－（1）－250 v． $70 \mathrm{~mA} . \mathrm{f}_{3} 3 \mathrm{v} \cdot 2 \mathrm{a} .5 \mathrm{y} .2 \mathrm{a} . .169$

 3n0－0－30）

 FIT．）sifto ind 1 \＆ Modrel tvpe $2-3-3 i n .2 a, ~ 176$

 40－0－50］ 100 m \＆，\％子 v

 for Mullay 510 A．mplifier （30） $0-550 \mathrm{v}, 150 \mathrm{~mA} .6 .35$ \＆ 4.5
 2A． 5 v． $3 x$ $425-92$ v． $200 \mathrm{~mA}, 6,3$ v． 4 a，C T Wlldanarom Amplifier，ele．
 All wilh $2 \pi-250$ v． 50 e s．primulies $t, 3$ v 1.5 w． $5,9: 6,3 \mathrm{v} 2$ a． $78 ; 0-4-6.3 \mathrm{v} 2 \mathrm{a}, 79$
 $12 \%$ ： 12 ， 3 ，or 24 v． $1.54 .1 \% 6$.

Brind nsw R．F．trot 17 6．A．F． 76
 diam．splndle all valnes lese switch． switeh． 46.

 1）utput 120 v． $40 \mathrm{~m} . \mathrm{A}$ ．Fully smoothed and －octified supply q．o charke 2v．aceumulator． Proce with lowved mretal hase and cireuat． 396 ，wrendy for uke， 89 extich．
 250 inA． 20 H 200 ohmis

198 250 mA .5 H 50 nhms
150 mA ． $6-10 \mathrm{H} 150$ ohm＇s Troip
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50 mA ．6－10 H 150 alims Trop． 69 120 mA .12 H 100 ohins

69
99 （0）mA．－ H 100 ohms

311

All with 200－2ibu－254 v． 50,1 Primatits
 $0-4-15$ v． 5 \＆． $199 ; 0-4-15$ v． 6 \＆． 239.

## S1世MVIIIVi THAKKE

$250 \mathrm{~mA} .5 \mathrm{H} 1(\mathrm{~K})$ whits
$150 \mathrm{~mA}, 7-10 \mathrm{H} 250 \mathrm{ohm}:$
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Small Pentode $78.0 \times 0$ se to 312
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Pu－h－Pull $10-12$ watte tiv6 wo $3 \Omega$ ur 150
 tors－a－8（br 168




 $\because .59 \mathrm{C}=375-5-55 \% 150 \mathrm{~mA}, 6.3$ v． 4 a， C．＇$\quad 6.3$ v 1 F．Filly inrouded． 229 1）rop lihrounh type． 189 ．Postage on ellhel Eype 23.

## R．S．C．BATTERY TO MAINS CONVERSION UNITS <br> Tyoe HM2 size ox $5 \frac{1}{2}$

Type BM1．An $\underset{+1}{ }$ b－dry bat er y eliminator．
 applat． neplathe 1.4 v．and $0 \%$ ． phyng 1．4 v．dand on
where A．C．mains 200 － where A．C．mis avasl－ able．Stifalib＊for atl thattiars of ry aly
 1．4 1．ans 90 \％．This impluden latert low consumption typer． Complete kit with（iatrams，39／9，ur ready t．s use． 46.9.
 21 in ．Supplies 180 v ．
90 v ．and $60 \mathrm{v} . \mathrm{m} 40 \mathrm{~mA}$ ． （9） y ．and 60 V ． 40 mA ． and 2 v． 0.4 a． 10,1 amp． fubly smont hed Therts－ lis＂orndletols rex lhating both 11. ．ing batlarios mind l．ar． When ronnected io A．C．meane supply

 EEEs normaitr using 2 v．acemmulator Eircs nom hair mate diagramsand inntructjons． 499 ．ur thady for use， 596.

HINI ITEKB．Movorsa． 2428 v．D．C．wr A．C．made hy llonver lat．．Camada．size oniv $2 \cdot x$ lin．Spindle lim．Iong，in． dam．Rrand NFw， 9.9.
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＋itn．2－3 6hms． 2911. Hith．2－3 whms． 359. Vers llmurd number

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4 mfd 350 v．． $298 ; 4 \mathrm{mfl} 1.000$ v． 49

UI．IVIV．NTAR TR INWHOHKMERA
Primaties 2（0） 250 v． 50 r．

159
99
 A design of a t－value lomg and Nedium wave $230-250$ A．C．Natins revelver with ＊elentum recifier．It eonsista on a high gain H F anode bend decectar．Power pentode rut－ put is used．Valif linewp bKT．Sibil． 5VbG：Selectisily and quality are wit up to standard．And simplic＇ity al conselturt tion is a sper 1 al frature．Point－ta－point wiring diatrams，instrue＇tions and part－ list： 19 ．Maximutur bulding mons－ £196．including sttrative Brown ol Cream Hakelite wr Walnut venecred wood canjnet $12 \times 41$ s 5 bin．

 $10-17-100-200-220-240$ v．to $5-0-75115-135$ 与 al RHVEHSE．HO－1u）watts．on ly 129. lus 29 post． $10-0-100-200-220-240$ F． 10 －0－110－122－136－148 V．Or REVERSE． 200 watts 359 ．blu：$\%$ 14arr．Both 50 e．p．s．
 frimarv 0－110－120－200－210－220－240－240－250 v． 0 －p．s．Sess． $275-(1-27 \mathrm{~s}$ v． 100 ma ． 6.3 V．． a，5， 3 d．Guvt．rating．22 9．Following with $243-230$ v．primaries， $400-0-400 \mathrm{v}$ ． 200 inA． 5 v．is \＆． 5 v． 2 a． $199: 230-07-280$ v． 110 mA .12 .6 v． 1.5 d． 3 v． 2 a． 119 ： 12.6 に． 3 d． 5 v． 3 \＆． 9 ．Postage 29 on any type．
 Wull ventilated，black＇ravkle Inizhed， Hid CHAR（IER ITR INS RUMENT CASE． OR COVER COUNO BE slze $81 \times 1: 5:$ Bij 12 s ．with undrilled wall ventilated oovel．finished in stoved yres enamel．subtable jor ，harger or thatument rase． $7 \cdot 9$ ，phus 29 post．

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| 1R5 | $7 / 9$ | 6.5 | 59 | ECCAI | 49 |
| 1＇4 | 79 | $6 \pm 6$ | 4.9 | EF80 | 7 |
| 185 | 79 | 6 V 6 c | 7.9 | $\mathrm{EBO}^{\circ}$ | 48 |
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| 65．J7GT | 619 | 15D2 | 49 | EZ90 | 0.6 |
| 6SIGT | $8 \cdot 9$ | 35\％4GT | 89 | E1．84 | $10 \cdot 6$ |
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| 6 AT6 | 79 | ECCB3 | 99 | SPbl | 2 |


|  i32－32－32 m｜d． 250 v ．Dubilier small cans．＇2 9 ea． 150 mtd .450 v．．3． 9 ．small .0005 mid． 2 －gang． 4,9 ea．Westing－ house Rectrfiers $250 \mathrm{v} .250 \mathrm{~mA}, 79$ <br>  8d．Yd．Twin－Screened Feedel 11d，yd． |
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16 mfd． 150 v． 1 16 wF 450 V ． … $150 .{ }^{2} 11$ 2 mld .450 v． 4.8 $\begin{array}{ll}100 . \mathrm{mfd} \\ 8-8 \mathrm{FF} & 450^{\circ} \text { V．．．．} \\ \text { V．} & 4.9\end{array}$ | $8-8 m F$ | 150 |  |
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| $8-16 \ldots \ldots$ | V．．．． | 2 | $16-16 \cdots \mathrm{~F} 450$ v． 411 $32-32 \ldots \mathrm{~F}$ Y 50.4 .9 $100-100$ mld． 351 ） 5.9 $100-100 \mathrm{mfd}$.

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275 v．．．． 68
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| $0-25 \mathrm{~V}$. | 0 | 250 V |
| 0- 100 V . | 0 | 1.000 V . |
| O- 250 V . |  |  |
| $0-1,000 \mathrm{~V}$. |  |  |
|  | 0 - | $100 \mu \mathrm{~A}$ |
|  | $0-$ | 1 mA |
| Resistance | C- | 10 mA |
| 0-20.000 ? | 0 | 100 mA |
| O-2M 12 | 0 | 1 A | List Price:


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#  <br> EVERY MONTH <br> VOL. XXXIV, NO. 61Ğ, JUNE 1953 <br> COMMENTS OF THE MONTH 

## MII) (iETS

Editorial and Advertisement Offices : PRAC"TRCAL WIHEAESS
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The Editor will be pieased to consider articles of a practical nuture. Such̀ articles showld be written on one side of the raper only: ard should comain ble mame and address of the winler. Whilat the E:ditor does nor hrold himhely risponsible for mamuscript. cvery. flomt will be made to return them is is stamped and coddressed covelope" is emolosed. All correspondence intended for the Fditor showld loe addressed

 Southampton Street. Strand. If.C.2 Uning to the rapid progress in the devigh of wireless apparatus ond ru onf pfforts to keep our readers in ton!! with the latess developments, we sine no warranty that apparatis descrithet it our coldmus is not thes subic: ot ienters pateht.

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I

OUR recent editorial comments on the need for a really midget battery-operated personal receiver, light in weight, and easily carried in the pocket, has brought forth a large amount ol correspondence and suggestions, and readers will be glad to know that we are now experimenting with a circuit which we feel will meet most of the conditions which have been named. It is obvious that high quality of reproduction cannot be expected, nor can the reader expect to receive a large number of stations. Such a receiver must essentially operate only on the medium-waveband and from low-voltage batteries. One or two readers have submitted tiny receivers which they have made themselves and we shall be describing one of these in an early issue. In the meantime. we shall still be glad to receive details of ${ }^{\circ}$ midget receivers which have actually been built and we should prefer constructors to send the receiver to us for test. We shall. of course, pay for all articles describing such receivers accepted lor publication. One of the difficulties at present is to find suitable sources of components, and it would seem that some of these will have to be made.

## - NOTHER "I'W'" EVIENING:

EVER since we organised that P.W. evening at Caxton Hall E last year there has been immense demand from readers for a repetition of it with, of course, a different programme. We should be glad. therefore, if readers who would be interested in attending will drop us a card signifying their intentions.

## THE RADIO SHOW

THIS year's Radio Show takes place at Earls Court from August 27th to September 6th. We can report up to the moment of going to press that there is no indication of any startling developments and the mixture it would seem will be as hefore. It is good that there should not be too frequent changes in design. The motor trade tried the sales dodge of putting each year's model out of date at the end of the current year. This was aluays irritating to the purchasers. The trade has now adopted a more sensible plan of producing new models by series numbers and not designating them by the year of introduction. The radio trade in its earlier years followed this system. but found that it did not work. The public did not too readily take to the idea of scrapping a $£ 50$ radio set because the fatest model had a different design of loudspeaker grille or a different tuning dial. New models every two or three years are tolerable. This years show. therefore, may show price changes and. as usual, different designs of cabinet. The public will go to the Radio Shou not only to see the manulacturers products, but also the sideshows put on by the Services, the BBC and other associations. The Radio Show is a fascinating event for this reason alone. There will be improved catering arrangements and innovations to add to the comfort of the visitor. -F. J. (.

[^2]
## Pound the Ulot Wireters

Broadcast Receiving Licences THE following statement shows the approximate number of Broadcast Receiving Licences in force at the end of February, 1958, in respect of receiving stations situated within the various Postal Regions of England, Wales, Scotland and Northern Ireland. The numbers include Licences issued to blind persons without payment.

| Region |  |  | Tor |
| :---: | :---: | :---: | :---: |
| London Posial.. |  |  | 1,047,961 |
| Home Counties |  |  | 1.056.672 |
| Midland |  |  | 781.325. |
| North Eastern |  |  | 1.016.681 |
| North Western |  | $\cdots$ | 758.308 |
| South Western |  |  | 654.548 |
| Wates and Border | Countie |  | 406.725 |
| Total England ard | Wales | ... | $5.722,220$ |
| Scotland |  |  | 759,319 |
| Northern Ireland |  |  | 180,774 |
| Grard Total | $\ldots$ | .. | 6.662 .313 |

During February the number of combined television and sound licences throughout Great Britain and Northern Ireland increased by 96.476 . bringing the total to 7.994 .723 . Sound only licences total $6.662,313$ including 330.238 for sets fitted in cars.

## Transistars in Explorer

COSMIC ray. meteorite and
temperature information
By "QUESTOR"
now relayed to earth from the globè-girdling Explorer satellite is being gathered with the help of many tiny silicon transisters made by Raytheon Manufacturing Company.
Extremely rugged and reliable, the pea-sized Raytheon transistors are in the satellite's telemetering circuit that sends coded data to receiving stations on earth.

The transistors. Type 2N328, developed under Army Signal Corps contract. are the PNP design. made by the Raytheonperfected fusion alloy process. This process permits a mechanically stronger assembly. allowing the transistor to withstand higher shock and vibration stress, both in the launching and orbit.

Other advantages for satellite use. in addition th the many inherent advantages of silicon transistors. are the $2 \mathrm{~N}, 328$ 's high electrical efficiency and ability to withstand sudden surges of voltage.


Mr. D. P. Young, Chief Installation Engineer, Marconi's, is seell receiring the congratulations of the Minister, Chief Akimola on the surceess/it opening of the network referred to on the opposite page.

The silicon devices work well lemperatures ranging from minus 50 to plus 150 degs. Centigrade, or much higher than germanium. Likewise. silicon performs more uniformly at the predicled temperatures for which the satellite equipment nas designed.

## Retirement

THE retirement is announced of Mr. W: H. Grinsted. O.B.E., F.C.G.I.. M.I.E:E..

wr. Girinstert, O.B.E., cuc.. who has retired from Ediswan as reported here.

Direcer of Engineering of Sientens Edison Swan I.td.

Mr. Grinsted was born al slinfold. Sussex. He receited his education at. Collyers School. Horsham and at the Cemral Technical College (now Imperial (ollege) London. where he held the Clothworkers scholarship and won the Siemens Medal fawarded to the third year siudent heading the list in Electrical Engineering).
R.E.C.M.F.'s New Chairman $A^{T}$ a meeting of the Council in London. Mr. K. G. Smith (N.S.F. Ltd.) was elected Chairman of the Radio and ElecIronic Component Manufacturers ${ }^{\circ}$ Federation in succession to Mr . Richard Arbib. Mr. Hector V. Slade (Garrard Engineering and Manufacturing Co. Ltd.) vas elected vice-chairman.

Mr. K. G. Smith is technical and sales director of N.S.F. Lid..

Whose factories are at Keighles. rorkshire and Liverpool. He is a mative of South Africa but has liced in this country fer one ${ }^{3}$ yars

He is also a director of the Motor and Electronics Corporation. I.td. of British Centralab. I.tu.

Lincar Accelerator at Cheshire Hostital

AFOUR million electron-toh gantry mounted lincar - Accelerator designed for supur whtage X-ray treatment ol deepseated tumours has been instatled in the new Radiotherapeutic Unit of the Clatterbridge (ieneral Hospital. Bebingtom (heshire.

The inaugural ceremony. pur formed by Lord Cohen of


Radio Industry Council. 'these were sound reproducing equipment. where the value in one month lopped fl 1 m . for the first lime. and valves and tubes. where the value was $\$ 420.000$. These ligures compare with valuen ol $\ddagger 743.000$ for sound reproducing cquipment in February, 1957.
by the Nigerian Miftister of (ommenications and Aviation. Chicf the Honotirable S. I. Naintola. M.H.R. Marconi's Wircless Telegraph Company hate ben responsible for the construclion of the whole system, which is capable of providing an ultimate capacity of over


March 28th under the auspices of the Liverpool Regional Hospital Board.
the installation of this equip. ment represents a further stage in the programme for the probision of Linear Accelerators for deep therapy treatment at suitable hospital centres in Cireal Britain. Newcastle-upon-Tyae Cieneral Hospital was the tirsi lo be equipped under the programme. Other centres where linear Accelerators have been installed are in Edinburgh. Iondon and Nanchester.

NewRecords for Radio Exports E XPORTS of two items ol radio equipment reached their highest monthly levels in Febraary it is announced by the
and of $£ 350.000$ for valves and tubes in February: 1957.

The provisional value of exports of all items of radio equipment in February was $£ 3.6 \div \mathrm{m}$. This was slightl: below the figure. $£ 3.7 \mathrm{~m}$. for February. 1457 (but may exceed it when the full returns are hoown

The total for the first two months of the year is now over $\$ 7.4 \mathrm{~m}$. compared with $£ 6.9 \mathrm{~m}$, for the tirst two months of 1957. which was a record year.

## World's Biggest Radio Multi

 channel LinkTHE largest radio telecommunications system of its ty pe in the world was inatngurated recently in Nigera

Ontvide and inside views of the $£ 300,000$ factory at Swindon, which is tw he devored to the prodmetion of transivors. It has mo windows is wiusproef and fully air-conditioned. It is owned be Senil Conductors, Lid.
50.000 channel miles. Marconi V.H.F. multi-channel equipment has been used at all the 14 terminal and 25 repeater stations.

With the completion of this scheme internal communications if Nigeria have made a significant advance. It is now possible to telephone from Lagos to many provincial towns whose local services have not previously been connected into a national network. The important centres now linked by first class trunk services include lagos. Ibadan. Benin. Onitsha. Enugu and Kaduna.

## Radios to Seven Countries

HUNGARY is to export a total of 14.000 radio sets to Cyprus. Yugoslavia. South Africa. Swerlen. Czechostorahia. Grece and Roumania this year from the Telephone Electrical Goods Factory. Budapest.

And railsay safety equipment worth $£ 600.000$ will be exported to (zechoslovakia and rugoslayia, reports the newspaper Népaharat.

## A Beginner's TOWSTMELCYIONAL TOURSE-7II

A NEW SERIES WRITTEN ESPECIALLY FOR<br>THE AMATEUR<br>By E. V. King

3.-CONVERTING THE 2-VALVER INTO A 3-VALVE SET

WHERE the lead from C14 goes through the grommet scrape it clean and solder on R10 which is then earthed to chassis (Fig. 16). Solder on a few feet of lighting flex for the speaker. Fix C16 across the primary of the speaker transformer, i.e. the tags already going to anode and H.T.

## Checking the Circuit

Again pencil out the theoretical circuit. Ink it in as each part is checked. Especially check the value of R9, beginners make many mistakes with the colour code. Then carry out a visual "round the clock" valve base check as follows: 1 to earth. 2 to earth via R9 and C11. 3 to TR2 and C16. 4 to pin 7.5 to earth. 6 to earth. 7 to pin 7 V2, TR2 and pin 4 V?. 8 to pin 8 V2.

## Testing the Detector With an Output Stage

The detector stage should not have been disturbed and should of course still be working. Connect everything up, note the mains is in the right way round and switch on. Note that both filaments light up. If not check the mains supply and filament circuits. Attach a good aerial ( 50 ft . will give good volume at this stage) to the positions on the PAZ already detailed. In Fig. 22 these places are "Y" or "Red " of the PHF2 coil near V2. If the aerial is short it is best attached directly to the red tag of the coil via a small 500 pF safety condenser.
Your receiver hill now receise the local stations at good speaker volume. There should


Fig. 16.-New wiring added when fitting $V 3$ (wiow of underside of "chassis).

only be a minimum of hum from the speaker and speech and music should be undistorted and clear.

If you do nol get reception proceed as follows. Unplug V3 leaving all wiring in place. Attach headphones as already detailed and verify that the detector stage is working. If not then you must have disurbed the detector stage wiring or shorted the H.T. supply to earth. If the detector stage is working replace V3. remove the top cap clip and touch the top cap with a finger. Loud 50 cycle hum should be heard from the speaker. if not then the output stage is faulty. Check the valves by interchange into the detector stage (using phones). If the valves are all right thoroughly check all V3 wiring and the value of components.
If the directions have been carefully followed no troubles will arise. A student of 15 buit onic of these receivers from these instructions and only made one error which he quickly put right. His error, for some reason or other. was to short out RII with a direct loop of wire between C12 and C13. The result. of course. was a terrific hum even when all top caps were in place!

Do not procecd to the next stage until your two valver is working perfectly as regards clarity and lowness of ham. There are some notes on hum at the end of this series.

## Adding an H.F. Stuge

If we fit an amplifier to increase the aerial signal then we shall get much more volume from the output stage. This volume is. of course limited. or distortion will occur on local stations. So the amplifier fitted must have a variable amplification factor. If the amplifier fitted is


Fig. 17.-. Additional wiring when V3 is fitted (top riew of chassis).
tuned like the detector stage we shall hate much increased selectivity. Stations near together on the dial will be separated with ease and distant ones received well provided fading troubles are not too bad (these are atmospheric in origin).

## Preliminary Alterations to the Detector Stage

Fix another coil. a Wearite PHF2 this time. in the position shown in Fig. 14. Make sure the red tag is facing the top cap of V2 as shown. Now remove the end of C7 which is attached to the red tag of PAZ and attach it instead to the red tag of the PHF2 just added. Shorten the leads if necessary. Earth tag X of PHF2 and earth tag $Z$. Any suitable tay may be attached to chassis for this purpose. Take a short lead


Fig. 20.-Side view of the receiver with this month's moctifications.
from the red tag of this coil to the other gang (C6). that is the gang nearest the front of the receiver. making sure that the lead goes to the fixed vanes and not the moving ones.

Now try out the detector and output stage as before. It should work as before when the aerial is placed on the red tag of the new PHF2 coil. It should work. but with much reduced volume. in position on tag Y. This is in order. Do no: proceed until the set is working with this new coil added. Any troubles which occur


Fig. 18.-Theoretical circuit of H.F. stage.
must of course be due to the firmed circtuit of coil and condenser and nothitig stser, for nothing else has been disturbed.

## Wiring the H.F. Stage (C'nderneath)

Refer to Fig. 20 and ti the volume control in the correct position. Fi another Mazda


Fig. 19.-Top view of H.t. and detector stages.
Octal base for V : in the same relative position as V2 and V3 (Figs. 1 and 211 .
The theoretical cireuit is given in Fig. 18 , compare it with the diagram of the complete threevalver in Fig. 22. The under view of the practical layout is given in Fig. 21 and the top view in Fig. 19. The front elevation of Fig. 20 may also help the reader with the wiring.
Here is a suggested plan to follow: Take pin 8 of $V 1$ to pin 8 V2. Take pin 7 on V1 to pin 7 on $V^{\prime 2}$. This valve is then supplied with H.T. and L.T. Earth pin 1 to nearby tag. Earth pins : and 6 to the other tag. Suspend Kl in air between pir : and any earth (i.e. pin 6 , hould do). Fix C3 likewise. Take pin 3 to the $\gamma$ tag of the PHF? (detector stage). Remove the carth lead on the $Z$ tag and instead take $Z$ to C8 (other half of double $8-8 \mu \mathrm{~F}$ ). Join C8 wo pirl 7 on 11 via R5. The H. 1 . supply is now decoupled. Tahe pin 4 to slider (centre tag) of the colume control and from there to carth (any suitable tag is fitted) via C2. One of the side tags of the volume control is carthed. the other is connected to junction R5 and C 8 vid R3 (more about this value later). Earth junction of R3 and volume control via C4 (this could be omitted in most cases "ithout any cffect). Check this with pencil and paper. and " round the clock"
again as follows: 1 to earth. 2 to earth via RI and C3. 3 to $Y$ of PHF2. 4 to R2 and C2. 5 to earth. 6. to earth. 7 to pin 7 V2. 8 to pin 8 V2. Now check for sure that your component values are correct.

## Wiring the H.F. Stage (On Top)

The PA2 is now used. This should already have tags $Z$ and $X$ to chassis, and the red to the fixed vanes of the nearby gang C5. Make sure these leads are still attached. Now take

## LIST OF COMPONENTS

## FOR OLTPUT STAGE (Fig. 15)

C1I-25/F 25s.w.
C14-. 01 ,"F
C16-.01.4F
R9- 339 ohms
R10-470k ohms
Tr2-Standard speaker transformer
V3-SP41 61
a lead direct to the top cap connector of the new valve Vi. The H.F. stage is now complete.

## Testing the Complete Receiver

Fix up all leats making sure mains is in correctly. Observe bhat all valves light up. If not verify if the val.ve is faulty by changing them round. or if the filament circuit is faulty check the circuit Trl. and pins 1 and 8 on each holder. Adrance R2 and signals should be heard. A long acrial is not required. Advance R2 fully. if the set starts to oscillate this is in order. Do not leave it doing so, retard R2 until the oscillations clear (if there were an". and tune the gang condenser. Stations should be heard though probably with interference for the time heing.

Trimming the Receiver
Have a look at the gang condensers and find the small auxiliary condensers which are called trimmers. A small screwdriver will be required to adjust these. Do them up tight and then undo


Fig. 21.-.View of the urderside wiring added to complete H.F. stage.
them both two complete turns. Now open the vanes and find a weakish station (i.e.. Luxembourg, if you are not too near the south coast). Use R2 control as necessary, but do not let the receiver oscillate: if it does reduce setting of $\mathbf{R 2}$ slightly. Now adjust carefully the two trimmers


Fig. 22.-The theoretical circuit with this month's additions.

one at a time to receise your wals station at mavimum volume. If the radio oscillates as you do this retard R2. With care you will get the station loudest at a certain setting of each trimmer and if each screw is moved the slightest cither way the signal should be fainter. If this condition is not obtained you must experiment with one trimmer done up more than the other in the first instance.

All trimming is best done on the acrial it is intended to use with the receiver as the setting of the trimmer on C 5 is affected by the capacity of the acrial. The author would like to repeat again that when finished the trimming should be such that a weak station on the high frequency end of the tuning is peaked with cach trimner.
(To be continued)

## LIST OF PARTS

Here is a complete list of parts, with alternatives and their function in the circuit.
$\mathbf{C 1}-500 \mathrm{pH}$ mica aerial series condenser fitted to safeguard someone fitting the acrial if the receiver were connected to mains the wrong way round. Incidentally, it will also save PAz burning out if the acrial hit the gutter. etc., under similar conditions.
C2-. $01 \mu \mathrm{~F}$ paper; 350 volts working. This is a screen decoupling condenser.
C3-.01 $\mu \mathrm{F}^{\mathrm{F}} 350 \mathrm{v}$. bias condenser.
C4-. $01 \mu \mathrm{~F} 350$ v. decoupliog condenser (optional, sec text).
C5 and C6-Twin gang 500 pF , each gang fitted with thimmers. Slow-motion drive an advantage. A three-gang may be used with the middle gang not used tone prototype will be seen in the photographs to have a three-gang in use). Insulated knob with sumken grub screws required. These condensers tune the two coil secondaries to the frequency required.
C7-100 or 120 pF mica or ceramic. This is the grid condenser which stops a charge on the grid of $\mathbf{V} 2$ leaking instantancously to earth via PHF2.
CB and $\mathrm{C} 9-\Lambda$ double $8-8 / \mathrm{F}$ electrolytic condenser. Any value will do which is greater than 8, i.e., a 32-32 $\mu \mathrm{F}$ will be just as good. The working voltage must be at least 250 v . and for safety ordet 350 v . condensers.
C8 decouples with R5 the H.T. for VI. and C9 decouples with R6 the II.T. supply for V2.
(10-. $1 / \mathrm{F}$ paper 350 v. working, screen decoupling for $V 2$.
C11-25/F (between 12 and 50 is suitable), 25 v. electrolytic condenser. Do not use one with a higher voltage rating. This keeps the bias on V3 grid steady no matter what current is passing across R9.
C12 and C13 Another double electrolytic. 16-16/ti or larger, say, 32-32 $/ \mathrm{F}, 450 \mathrm{v}$. working. Do not use a lower working voltage than 350 v . C12 is the rescrunir condenser for MR1 and C13 is the smoothing condenser working wilh R11.
C1t-. 01 /F mica or ceramic or Sprague type, 600 v. working. This couples the output from V2 anode to the grid of V3. It inust be good or H.T. will get to the grid and ruin V3. Slight leakage would put a wrong bias on V3 grid and cause distortion.
C15-100 $\mathrm{pF}^{-}$mica or ceramic. This gives a H.F. by-pass to carth with practically no path for audio currents. It also helps with modulation hum troubles.
C16-.01/4 paper 450 v. working. A top cut capacitor for toning down the high note response of pentodes. It also helps to cut out medium wave heterodyne whistles during the evenings.
(. $\mathrm{N} \|$ resistors are I watt unless stated otherwise.)

R1- 1,000 ohms (brown, black, red). This gives a bias on V1 grid. Bias varies with position of R2 as the current through the resistor thus varies.
R2-1 megohm potentiometer. This is a irue potentiometer between H.T. plus and minus. The voltage
(and hence the amplification of the valve) on VI screen grid is thus controiled by the slider.
R3- 220 k . (red, red, yellow), a resistor to stop the grid of V1 connected to R2 slider from becoming too positive. The value given is satisfactory, for greater sensitivity adjust as described later.
K4-1 megohm, (brown. black, green). Girid leak to allow the charge on the control grid of $\sqrt{2} 2$ to leak slowly to earth, otherwise the valve would cut off.
R5-10 k. (brown, black. orange). Decoupling resistor for V1, works in conjunction with C8.
R6- 10 k . Decoupling resistor for V2, works with C9.
R7-1 megohm. Screen H.T. dropper for V2 to get a suitable voltage on V2 grid for leaky grid detection.
R8- 470 k . (yellow, violet, yellow). Anode load for V 2. The detected signal is developed across this recistor.
$\mathbf{R}^{9}-\mathbf{3 3 0}$ ohms (orange, orange, brown). Bias resistor for V3. This gives a voltage drop when the valve draws current, thus producing a negative bias for V 3 grid (via R10).
R10-470 k. (ycllow, violet, yellow). Grid leak to give the proper bias to control grid of 13 and to stop it acquiring a charge.
R11- $\mathbf{3 , 0 0 0}$ ohms ( $\mathbf{5 , 0 0 0}$ ohms would do). 3 watts or larger vattage. This is the smoothing resistor in the H.T. circuit and works in partnership with C12 13. Large resistors are not usually coded, but if so it is orange, black, red.
1,1-Wearite PA2 acrial coupling coil and tuned grid coil for V1.
1.2-Wearite PHH2 anode coupling coil and tuned detector coll for 12.
VI-SP41 SP61 (that is VR65a;'R65 ex-government) H.F. amplifier.

V2-Ditto. Leaky grid detector.
V3-Ditto. L.F. output stage.
Three bases for the above valves. Mazda Octal. Three top caps to suit.
TRI-Filament transformer, 4 v. 3A. or $6 \mathrm{v}$.2 A . isce text) from A.C. mains. Standard speaker transformer will do under some conditions.
TR2--Standard speaker transformer or multi-ratio type (R.C.S.). This provides an anode load for V/3 and matches the valic to the speaker.
M1R1-Metal rectifier. Any type supplying 30 mA or more and rated at $2+40 \mathrm{v}$. input will do. This rectifies the A.C. input to pulsating D.C., which is riceived by (12 (II.T.48 is a suitable type).
SI-Mains on/off switch. Toggle typȩ not suitable. Rotary type with insulated knob is required. One pole one way will do. or two pole one way can be used (see tevt).
CHASSIS-Sce text.
MAINS LEADS-Two core coloured cable with threc pin plug, if possible.
AERIAL- 50 ft . or more for one-valver, 20 ft , or less for ordinary reception with 3 yalves.

# New Radio Components 

detalls of some of the exhibits seen at this Year's r.e.c.m.f. ShOW

MANY of the new radio components which were exhibited at the Radio and Electronic Component Show at Grosvenor House and Park Lane House, London, were smaller. more robust and able to withstand higher temperatures than ever before.

The trend to miniaturise components continues. particularly for those to be used with the very small transistors now being manufactured, and also in connection with printed circuitry.

Components for use in guided weapons need to be sub-miniatire. of extreme ruggedness and able to operate in high temperatures. Some of the new transformers and chokes are suitable for working in temperatures up to 200 deg. C. and even 250 deg . C. and there is an exploratory design suitable for 500 deg. C.

Certain components have also been designed to withstand the strong vibration and enormous acceleration due to high " $G$ " values in guided missiles.

Improvements are 10 be found in even the more stereotyped components such as wafer switches and small relays. both sealed and unsealed.

Developments in ferrites and dust iron cores include the introduction of new materials with new properties, especially for memory devices for computers.

Made for the first time in this country are sintered glass preforms for glass-to-metal seals in hermetically sealed components. These are cheaper. less laborious to use and available with closer iolerances than the glass tubing they replace.

It would not be possible to deal with everything shown at the exhibition. so we pick out here a few of the more interesting items from the point of view of the constructor.

## Travelling Wave Tube

The T.W.S.1. shown on page 271 is a convection-cooled travelling wave tube by the



The New Series 100 Test Set by Pullin.
M.O. Valve Co.. intended for use in the 1.500 to $3,000 \mathrm{Mc} / \mathrm{s}$ frequency lange. and suitable for most broad-band amplifier and variable attenuator requirements. The tube is conservatively rated. and an output of 38 can be obtained with a power gain of 26 db .

## Series 100 Test Set

The Pullen (Measuring Instruments Lid.) test set shown above has been redesigned and has the following fcatures:
(i) A new type diakon plastic cover is fitted to the meter. which gives a very wide angle of vision. and good illumination on the dial surface.
(ii) An increased scale length is possible due to the use of the new cover. and re-arrangement of scales, giving an extra long ohms scale.
(iii) The test set is fitted with a stand which makes it possible to use an instrument in a horizontal. vertical or an inclined position. This, it

will be appreciated facilitates reading the instrument under difficult lighting conditions. The case has been re-styled in aluminium alloy and a finger recess for carsing is fitted instead of the protruding handle previously used.
(ii) The case is finished grey hammer enamel and an adjustable support is incorporated


Whiteley Electrical
Amongst the wellknown Whiteley (W.B.) apparatus were the 10in. High Fidelity Loudspeaker (1) pe HF.1016). incorporating powerfut 16.000 gauss magnet. cambric cone and lapped speech coil. handing capacity 10 watts. Irequency response 30-15.000 c.p.s.. retail price $£ 8$ including tas.
Also there was a new model F.M. tuncr with a circuit of advanced design. An important feature is the use of printed circuit techniques to ensure mavimum clectricity stability together with rigid mechanical structure thus ensuring freedom
(v) The meter has a 3 in mean scale length. rectangular pallern with hnife edge pointer. 3 ohm scale and a sensitivity of 100 microamps for full scale deflection.
(si.) The test leads have been improved and thes are fitled with retractable probes and a swivelling mechanism which enables the leads to be used in ans position without rish of damage.
(iii) A special printed cirenit has been designed for the majority of the wiring. which gives a much more rugged construction. facilitates service. and enables all components to be accessible when the back cover is remoued.

Whitele
V.H.F:IF. tmisir.

Collaro Ttiss and a ceramic pick-mp.



The Travelling Wave Fule by M.O. Valve (o.
from drift. It is designed to receive rirquency modulation signals in the band of 88-108 Mcis and is therefore suitable for use in most parts of the world.
Plessey IP.M. Tuner
Another luncr was shown by the Plessey company. This is for A.C. and A.C.ID.C. operation. using a double triode valve as R.F amplifier and mixer-oscillator. The frequency coverage of this unit is 87 to $10 \mathrm{i} \mathrm{Mc} / \mathrm{s}$.

## Collaro Products

Collaro Ltd. presented their new improved Tape Transcriptor Mark IV. which incorporates several modifications to the previous design. Among the nen features are the interconnected micro switch and fly-wheel brake. By means of these improvements both motors will be switched oil and the 113 wheel instantancously stopped when the "stop" buttor is operated. Any one of the sis smaller hoob will then start the machine in either direction without subjecting the motors to cucessive loading.

THIS receiver was designed to be used by anyone, anywhere, at any time. It is a four-valie superhet using coonomy valies and readily obtainable cheap batteries. For the above reasons the set was built around an Ever-Ready B. 12690 volt H.T. battery and the popular $12894 \frac{1}{2}$ volt pocket-lamp battery ( $7 \frac{1}{2}$ volt L.T. batteries are notoriously expensive). The valves used are DK96. frequency changer; DF96. intermediate frequency amplifier: DAF96, detector. A.V.C. and A.F. amplifier; DL96. output pentode; a contact cooled metal rectifier is used for the mains supply together with a VR105/30 voltage stabiliser. Since the main circuit details are reasonably conventional. and therefore the power supply and switching are of more interest the receiver will be dealt with theoretically in two sections. although full practical details will be given.

The power supply section will be dealt with first and the circuit is given on page 274.

## Power Supply

The power from the mains is rectified by MR. C8A being the reservoir condenser. R11 the dropper resistor, C88 and the VR105/30 smoothing and stabilising the output at 105 volts. Current from this supply flows through R9. the filament dropping resistor. and $A / 3$. a $500 \Omega$ surplus relay, finally returning through a Brimistor type CZ10 and the filament chain if the set is switched on. The CZ10 is included to absorb the filament current, 25 mA . when the set is switched off. the voltage across it then being about 15 volts. With the set running, the voltage across it drops to 9.05 volts, and at this voltage it only passes about 2 mA . The current flowing through A/3 causes it to operate swi/ching AI, 2 and 3 to the mains position. Pin 7 of the DL 96 receives the mains supply of 9.05 volts which flows right through to pin 1 of the DK96 and the chassis. From there if $S 2$ (part of the on/off switch) is closed it flows through R6. which provides part of the grid bias for the output valve. and back to the mains. About 1.1 rolts is developed across R6 since the H.T. and L.T currents flow through it on the mains position; also about $5 \frac{1}{2}$ volts are dropped across R4 and the first three valves. thus about $6 \frac{1}{2}$ volts bias is prisent for the output

> An A.C.|D.C. Set With Stabilised Power Supply Enabling it to be Plugged in Anywhere Without Adjustment Where 200-250 Volt Mains Exist. Also Containing Automatic Switch-over from: Mains to Battery if the Power is Cut Off. By G. Keating
half feeding the remaining valves in exactly same way that the mains supply did. Onli the output valve changes its working condition in that all the bias is now produced by R6. R13 and R7 carrying the H.T. current. R3 is used to by-pass the cathode current of the DL96 on mains supply, since the anode and screen currents must flow out of the filament and would, unless bypassed. overload the other valves. Similarly $k S$ by-passes the current from the DF96. The DAF96 cathode current is only about $0.1 \mathrm{~m} \Lambda$ and may safely be neglected. R2 by-passes the cathode current of the first section of the DL96 and the current through R1 on mains. On battery Ri and R2 are in parallel and bypass the cathode current of the second section which, it will be remembered, comes first on battery operation, because the current flow through the filament of this valve is reversed. As a result of these resistors the voltage is maintained at 1.3 volts per section on mains operation and is no greater than 1.5 volts per section on battery which is the operating conditions recommended by the manufacturers. The stabilised power supply ensures that the operating woltages are always correct whenever the set is operated from A.C./D.C. mains between 200 and 250 wols. If the supply voltage is below 200 volts at peak demand times then the stabiliser may flicher on and off, in which case R1I should be reduced in value to, say. 2 Ksz (a 15 k 1 w resistor connccted in parallel. will achieve this temporarily). This can also be used when the rectitier falls in efficiency after very prolonged use. The relay deserves special mention. it is of 500 s resistance and it closes at about $12 \cdot \mathrm{~mA}$. so the 25 m .1 filament current is more than sufficient for reliable
operation. The original had a two-pole changeover switch only fitted. so a bolt was mounted on the chassis in such a position that the armature just made contact with it when the relay was de-energised. As the mounting bracket was already insulated from the frame a wire from the frame to the negative terminal of the
for simplicity. Consequently it is wiser to wire in all components in Fig. 1 first. before wiring in the additional ones in Fig. 2. The signal is picked up by a dual wave Ferrite rod aerial tuned by C9A and applied to G 3 of V1. The oscillator section of this valve is a conventional series fed, grid


Fig. 2.-The receiver circuit. The mains section is shown on page 274.
L.T. battery provided contact A3. Of course. if a relay with a three-pole change-over switch was obtained. this modification would not be necessary. Any relay will do providing it will operate Ermily and reliably on 25 mA . The value of R 9 should be altered so that $\mathrm{K}^{4}$ and the relay A/3 total 3.4 K゙』 resis:ance.

## The Circuit

The remainder of the set will now be desscribed and Fig. 2 givesthe circuit diagram of this part. If should be noticed that most. of the components already detailed in the power supply section will be omitted from Fig. 2

## COMPONENTS FOR FIGURE 2 R17-27K $\Omega$

* C5A-50/F 150 v. ). R18-10 M $\Omega$
$\mathrm{C} 5 \mathrm{~B}-50, \mathrm{~F} 150 \mathrm{v}$. in the same can $\mathrm{R} 19-39 \mathrm{~K} \Omega$
C9A, $500 \mathrm{pF}^{\mathrm{F}}$ twingans midget R20-47 K $\Omega$
C10- 5 pF silser mica or ceramic switch)*
C11-0.1 M 150 v.
C12-0.1 $\mu \mathrm{F} 850 \%$.
C13-100 pF silver mica
C1 -470 pF silver mica
C15-150 pF silver mica
C16-0.1 11 $150 v$.
C17-0.1 $1 / \mathrm{F} 150$ v.
C18-0.1/IF 150 v.
C19-100 pF ceramic
C20-100 pF ceramic
C2t-10.000 pF ceramic
C22-10.1/4 150 v
C23-10.000 pF ceramic
R12-10 1 !
*R13-100!
R14-27K:
R15-100 K!
R16-10 M!

C9B 1 with trimmers ( $2 \frac{1}{2} \times 1 \frac{1}{2} \times 2 i n$.) R21-2M $\Omega$ volume control (with double pole

R22-10 M $\Omega$
R23-2.2M $\Omega$
R24-10 M $\Omega$
R25-15 MI
R26-10 M $\Omega$
T1 and T2-Pair slug tuned midget IfTs, 13/16 x $13 / 16 \times 2 \mathrm{in}$. maximum.
73-Radiospares miniature output transformer
L1 Dual wave Ferrite rod aerial (Teletron used in L2 original but any should do). $8 \times 5$, 16 in .
L3-Osmor QO8

## L4-Osmor Q09

S3.A
S3H
L.S.- 3 ! loudspeaker (original was 5 in. Celestion but a $7 \times 4 \mathrm{in}$. elliptical would suit). $\Lambda$ high ${ }_{*}^{\text {flux }}$ deensity type is strongly recommended. *See Fig. I.
tuned circuit with separate coils for medium and and long waves.

It will be observed that both the oscillator coupling windings are in series with the mediumwave section nearest to $\mathbf{G 2}$, the oscillator anode.

Because of this the coils should not be mounted too close together as mutual coupling will make adjustment of one coil core affect the other. With the layouts given. however, the coils are virtually completely independent of each other. (14 and C15 are the medium and long wave padding condensers: these particularly must be silver mica. It is desirable that all the condensers marked in
to the I.F. amplifier $V 2$ and then through $T$ to the detector. V3. It is rectified by the diode section and the audio voltage. having been freed from I.F. voltages by the ilter R20. ('14 and (20), is developed across the volume control R21. A ? M! component was used in the original but 1 Ms! would be quite satisfactory if dasier to obtain. The audio voltage is taken from


Fig. 1.-The power supply circuit.
the components list should be silver mica also. The reason for this is that ceramic condensers have a fairly large negative temperature coefficient. and since the set gets fairly warm when used from the mains these will. cause scrious and prolonged tuning drift. You have been warned! Moulded mica condensers are mostly unsatisfactory, solely because of their size. To continue with the, circuit; the signal is now fed. through Tl the first I.F. transformer at $465 \mathrm{kc} / \mathrm{s}$,
the slidei of R21 and fed through C21 to Ci, the signal grid of V3. R22 is the grid leak, and R23 the anode load. From the anode of V3 the signal is fed through C23 to the output value V4. The output transformer matches to a 38 speaker and R25 provides sufficient negative fecdback to reduce the distortion produced in the output valve. The tone has been found to be quite good although the high $Q$ coils. which give cxcellent selectivity, naturally reduce the

## COMPONENTS FOR FIGURE 1

R1-680
R2-2.2 K !
R3-2.7 K $\Omega$
R4-56!?
R5-680 2
R6-33!
R7-560!
R8-2.2 K $\Omega$
$\dagger 129-2.5 \mathrm{~K} \Omega 5 \mathrm{w}$. wire wound $+390 \Omega \frac{1}{3} \mathrm{w}$.
R10-CZ10 Brimistor
R11-2.5K 910 w . wire wound
R13-100 $\varrho$
C1-25 //F 25 v. (BEC, CE200)
C2-25 / F $25 \%$ (BEC, CE200)
$\mathrm{C} 3-0.1 \mu \mathrm{~F} 150 \%$.
C4-0.1 $\mu \mathrm{F} 150 \mathrm{v}$.


C7- $25 \mu \mathrm{~F} 25 \mathrm{v}$. (BEC, CE200)
$\left.\begin{array}{l}\text { C8A- } 8 \mu \mathrm{~F} 275 \mathrm{v} . \\ \mathrm{CB}-8 \mu \mathrm{~F} \\ 275 \\ \mathrm{v} .\end{array}\right\}$ in the same can

C24-0.1 $\mu \mathrm{F} 150 \mathrm{v}$.
S1 and S2 ganged with volume control (R21)*
A/3 500 ? relay (see text)
MR Electrix contact cooled 250 v. 85 mA hall wave metal rectifier

+ Two resistors in series
* See Fig. 2.


## SUNDRY COMPONENTS

## 4 B7G valve holders

Octal valveholder
1.J.B. cord drive, standard type No. 4,690

1 .J.B. dia! drum, 2 in., No. $\mathbf{4 , 0 3 0}$
1 cord tensioning spring, No. 4,587
2 tool clips
1 plug to suit B.126, 3 pin
Various nuts and bolts (mostly 6 B.A.)
2 knobs to taste (original were A713 isin. White, engraved "Vol/On-Off" and "Tuning")
1 sheet of aluminium, $8 \times 13 i n$. for main chassis
Misceilaneous picces for screens, mounting brackels, etc.
Wire, solder, nylon drive cord, small pulley
treble response. As some people may prefer a further reduction in treble (what is usually called a mellow tone). it will not be amiss to show how this can be achieved. R25 the negative feedback resistor is omitted and a fixed condenser is wired across the primary of the output transformer. Another 10.000 pF ceramic
auviliary outside aerial is used. and to minimise variations in volume, caused by the marked directive properties of a Ferrite rod aerial. when the set is being carried as a portable. The negative bias point referred to carlier is obtained from the bias supply for the output valve and is arranged to be the same on both battery and


Fig. 3.-Plan larout of the portable.
will probably be found suitable. although the value may be varied to suit individual taste. Since the bass response can be surprisingly good for this type of set. a Sin. loudspeaker of good quality is recommended. A high flux density speaker has the supreme advantage of giving more volume from the limited output power. and is strongly recommended. Cheap and midget speakers will be found disappointing from the p गint of view of both tone and volume. The A.V.C. system is a little unusual in that it consists of a potentiometer network between the positive filament pin of V3. and a negative bias point of approximately -1.2 volts to chassis Three 10 MSE resistors R18. R16 and R12 divide the majority of the standing voltage of 5.1 volts into three cyual parts (the volume control disides a little but the amount is small). Thus at the junction of R12 and R16 the potential is about +0.5 volts to chassis and at the junction of R16 and R18 the potential is about +2.1 volts. 7 hese points are thus at about the same voltage as the middle of the filaments of $V 1$ and $V$ ? respectively and are thus suitable for the A.V.C. feed. Owing to this potentiometer, only one-third of the control bias is applied to VI and only two-thirds to V2. Another useful feature is that about I volt delay exists owing to the biases being about mid filament potential and not negative tilament potential. This stops weak signals decreasing the gain and improves sensitivity. A.V.C. is provided to stop strong signals from overloading the set if an mains. (On mains it is the full mains bias. on battery it is just part of the full bias.)

## Constractional Details

As every theoretical detail has been exhaustively deaft with the practical layout and dimensions will now be given. Despite all the refinements incorporated, it is remarkably compact: the measurements of the chassis being 8 in . $\times 6$ in. $X$ 5in.. which includes both batterios and a sin. loudspeaker: in fact everything except the cabinet and knobs. The chassis is made from sheet aluminium. as ready-made chassis are costly. and the wrong shape. The front panel forms the basis of the set and carries almost all of the components. Slow-motion tuning and a dial are included.

No detailed layout diagram can be given owing to the extremely compact wiring. but the layout of all the larger components such as valces. tuning condensers. transformers. and controls will be given.
(To bé continated)


A general view of the set.


## "Arrangements"

HAVE you ever noticed the high proportion of "artangements" pumped into the cther by ersatz musicians today? A composer writes a piece of music and immediately someone wishes to "arrange" it, so that it sounds nothing like what the composer intended. Some of those who do the arranging are not genuine musicians and could not compose music. Music should be played as the composer wrote it. So many of these quack musicians batten on like limpets and barnacles to other pcoples work The arrangement usually consists of a complete mutilation of the score so that the original theme is unrecognisable. 'Liszt was one of the pioneers of the arranging racke*. He arranged a large number of other people's compositions. This is a form of musical piracy. If so-called musicians feel they are able to arrange, let them compose and arrange their own music.

The Voice of América
THE VOICE OF AMERICA HAM SHOW is heard weekly on Tuesdays at 4 p.m. Eastern Standard Time ( 2100 G.M.T.) and is directed to audiences behind the Iron Curtain. It is a 12 minute segment of the regular English language 30 -minute programme called "Report from America" which is broadcast daily. As readers know, The Voice of America Programmes in Russian and other Iron Curtain languages are severely jammed, but broadcasts in. English are free from jamming. Since English is the second language of a substantial part of the audience which The Voice of America is trying to reach, these broadcasts are regarded by the American authorities as extremely important. The commentator for Voice of America is Bill Leonard.

## Three Exhihitions in One Week

INN April, no less than three comparatively small exhibitions were held on one week-the Radio Components Show, the London Audio Fair, and the Instruments, Electronics and Automation Exhibition. None of the industries which these three exhibitions represent is large enough to promote an exhibition on a national scale, with the possible exception of the last which held its show at the Olympia. The Radio Components Show was held in two parts at Grosvenor House and Park Lane House, London, and the London Audio Fair at the Waldorf Hotel. I s:ill think that these exhibitions should be held under the same roof as the National Radio Exhibition at Earls Court. Each of the three industries represented are only component parts of a major industry. I know that they are more or less trade slows, but that is no reason why the public should not be permitted to examine the exhibits.

## Luxembourg and Religion

I OBSERVE that the perfervid tub thumping, religious programmes are still being injected into the ether for five nights a week on 208 metres between the hours of 11 p.m. and midnight. There are usually four programmes of a quarter an hour every evening, and in all of them we are told what miserable sinners we are, backed up by lashings of miscellaneous and unrelated quotations from the Bible. All of these programmes emanate from: America, where religion has become big business. They do not teach it-they sell it. The thinly-veiled appeals for sacrificial letters from listeners must bring in a large amount of money. You: are offered frec copies of periodicals and booklets. Onc such periodical is before me now. $\mathbf{I n}_{\mathrm{i}}$ it is a column on how to make your will. It "includes the line ." to the ..... organisation" just to make quite sure that you do not forget them when you die. There is a religious mania in America. and there are blatant advertisements in the newspapers. One religious quack advertises "For one dollar you will receive a guaranteed blessing uithin 48 hours." Steps are, I am glad to note. being taken to investigate some of these religious organisations, many of which are run by people who have had no religious training. Unfortunately, under the American constitution laid doun by Abraham Lincoln, no law can be made which restricts religion in any way. Of course, Radio Luxembourg is a commercial station and the money of a religious organisation is as good as anybody else's. I recommend you to listen to some of these broadcasts and see if you agree with me. They do not ring true.

In America anyone can set himself up as an evangelist and millions of unthinking peóple may regard him as a trained parson. Some of these people are undoubtedly just making á business of it. and it seems a pity that they are aided in their efforts by being permitted to purchase programme time. Fortunately, it could not happen over here, but the ether knows no boundaries and large numbers of people do listen in to these foreign broadcasts. It is a thousand pities that our own Archbishop of Canterbury should have lent support to this form of insincere hot-gospelling and, in view of the exposures of this religious quackery which have taken place in the American papers, I hope he is wiser now.

[^3]
## Speaking for BRITAIN

What does Operation Britain mean to us?

at home and abroad, too little is known or even remembered of the active part that Britain plays today. We feel it. We believe in it. But is this enough?

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all parts of British life-experts in their own field of activite: labour, management, govern-mont-united by their belief in Britain's future.

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Our, is a heritage no longer bounded by the seas: ours is a vital role in world affairs. It is only by knowing this, cach onte of us, aware of our real ability, that we can build towards the future that is our birthright-the world's birthright-a future that is bright with promise.

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ADDRESSS.


Whereas in a multistage valve amplifier all stages right up to the output stage require only low dissipation circuits (assuming Class A working; the driver stage of a Class $B$ valve amplifier is more like the trunsistor case) a multistage transistor amplifier can be expected to be designed for successively greater dissipation stage by stage towards the output. This was seen in the two-stage audio amplifying circuit described in this series. The present purpose, however, is to deal with the final stage of the transistor amplifier-that which has the job of driving the loudspeaker.

As with valves transistors can be operated in Class A. in which ease the base bias is set so that the quiescent operating conditions (i.e., without a signal applied) are miduay along the straight part of the collector characteristic. or in Class B in which case the transistor is biased back so that the quiescent state collector curtent is negligibly small. The Class A condition gives amplification with comparatively little distortion and either single output transistors or push-pull circuits can be operated in Classs $A$. Class B. on the other hand, is fundamentally a distorting amplifier-in fact, a theoretically perfect Class B anplifier (if such were-possible) would amplify only one half-cycle and would cut off the other halt-cycle-and so for the purposes of audio operation two Class B transistors must be operated in push-pull, each amplifying one half-cycle only, but when the two outputs are combined in the output transformer there is. in effect, an amplified signal following both half-cycles of the input signat.

## Class A

The maximum theoretical efficiency of a Class A stage in the case of either valves or transistors, is 50 per cent. : that means that no more than halif the power drawn from the valve H.T. or transistor battery supply can be converted into an audio signal and any attempt to increase the dive beyond this


Fig. 39.-Low whage collector characteristic-OC72.
point merely introduces distortion without increase of output. This may not seen so good, but in practice even suich efficiency cannot be attained. The maximum of 50 per cent. efficiency involves three impossible assumptions :

1. That there is no loss in the collector load, i.e., that the voltage drop in the load under D.C. conditions is nil. This requires a transformer having a primary that has sufficient inductance for the purposes of loading the transistor but with zero D.C. resistance. This is impossible, of course, but fortunately the primary inductance needed by a


Fig. 40.-Dissipation characteristic-OC72.
transistor for a given frequency range is comparatively small and so it can be wound in a small volume and with low resistance.
2. That the transistor can be operated down to zero collector current. This cannot be so, either. There is a minimum level below which collector current cannot fall and this level is subject to variation with junction temperature ; this minimum has been referred to before in this series and is known as ic $(0)$.
3. That the transistor can be operated with a collector swing down to zero volts. An attempt to do this could run into gross nonlinearity, and consequently severe distortion, because the transistor current gain falls off rapidly as collector voltages fall at the lower end of the characteristic.

Fortunately, with regard to 2 and 3, the transistor can operate down to quite low collector currents and voltages and so, in fact, efficiencies reasonably close to the theoretical maximum can be achieved.

There are two possible approaches to the design of a Class A amplifier. Either the need is for the maximum power output that can be obtained from a small transistor of limited permissible dissipation or else the transistor is more liberally rated for the purpose in hand, in which case the choice of power output rests with the designer who can determine the power by the circuit constants. The object of the exercise, in both cases, is to adjust the base bias current until the appropriate collector current is obtained.
Taking the idealised case of a transistor in Class A giving the theoretical optimum efficiency of 50 per cent., and tearing in mind that the maximum dissipation of power in a Class A circuit takes place in the no signal, or quiescent, condition, it follows from the definition that the power output obtained will te a half of that dissipation. So, if the intention is to get the maximum audio from a given transistor, signal conditions can be put aside temporarily and
the D.C. conditions computed from the standard formula for power, i.e.:

$$
\mathrm{W}=1 . \mathrm{E} .
$$

where $W$ is the maximum dissipation, $l$ is the collector current Ic and $E$ is the applied battery vollage. The unknown is, of course, ic so by manipulation of the equation :

$$
\mathrm{Ic}=\frac{\mathrm{W}}{\mathrm{E}}
$$

If, on the other hand, the limiting factor is the power needed and not the permissible maximum dissipation, the first thing to do is to convert this audio power figure into a dissipation figure, which as above simply mears multiplying by two. Then the same formula applies, i.e.

$$
\mathrm{lc}=\frac{2 P_{\mathrm{L}}}{\mathrm{E}}
$$

Pl being the maximum audio power required in the load resistance.

In practice, allowance has to be made for the deviations from theoretical maximum efficiency. This has the effect of reducing the audio power that can be drawn as has been stated, due to the fact that the collector cannot swing to zero volis. This requires that the quiescent collector current must be set somewhat higher than was previously indicated if the transistor is to operate about the middle of its characteristic and the new ic is found by reducing the factor $E$ used in the above equation by an amount equivalent to the forbidden "territory" of collector voltage caused by curvature of the characteristic. It is wise to err on the safe side here, so commonly a half a volt has to be allowed for. Thus the equation for quiescent collector current becomes

$$
\mathrm{Ic}=\frac{2 \mathrm{PL}}{\mathrm{E}-0.5}
$$

This inercase in Ic is strictly necessary only if it is intended to operate the transistor over the whole of the straight part of the characteristic. If the transistor is to be operated well within its rating it is not likely to swing so far as the curved regions and so the adjusiment reed not be considercd.

This does not really help the man aiming at the maximum power from his transistor, however. He must still begin his calculation from the maximum, dissipation point of view which, of course, still involves the total wolage applied so again he must use

$$
\mathrm{Ic}=\frac{\mathrm{W}}{\mathrm{E}}
$$

By equating these two alternative hays of defining Ic it can be seen to what extent power oultput in practice falls short of 50 percent., i.e.,

$$
\begin{gathered}
\frac{2 P L}{E-0.5}=\frac{W}{E}(=I c) \\
2 P L E=W(E-0.5 W) \\
P \mathrm{PL}=W \frac{(E-0.5)}{2 E} \\
=\frac{(E-0.5)}{E} 50 \% \text { of } W .
\end{gathered}
$$

The effective drop in efficiercy from the theoretical maximum is thus less significant the higher the voltage applied and so it is clearly tetter to ircrease the audio power output taken from a transistor by inercasing the voltage applicd rather than by increasing the current drawn from the same voltage source.

Let us try some examples:-
If a $1 \frac{1}{2}$ volt source were to be wed the cfinciency
when maximum power permissible is being drawn from a single transistor in Class $A$ will be:-

$$
P_{L}=\frac{1}{1.5} \quad 50^{\circ}{ }_{0}=33^{\circ} \text { of } \mathrm{W}
$$

If a 6 volt system is in mind, such as the power supply used for the receiver developed in this series. or a car battery in the 6 volt range

$$
\mathrm{P}_{\mathrm{L}}=\frac{5.5}{6} \quad 50^{\circ}-46^{\circ} \text { or } \mathrm{W} .
$$

A 12 volt car system would provide only slightly higher efficiency though, of course, the current drawn would be halved because the voltage is doubled for the same dissipation. Actually

$$
\mathrm{P}_{1}-\frac{11.5}{12} \quad 50_{0}^{\circ}=48^{\circ}, \text { of W. }
$$

Abowe $\frac{1}{2}$-volt was quoted as a safe limit of swing of the collector voltage, and to support this Fig. "3y gives the collector characteristic of the OC72 at low voltages. It will be seen that this limit errs on the side ol salety so far as this transistor is concerned. Fig. 40 is given to show how dissipation varies with ambient temperatare. It will be seen that at 75 deg. C. no collector current can be allowed because this is the limit of junction temperature and any dissipation would raise the junction temperatare higher than ambient temperature. When ambient temperature is lower. the permissible dissipation is that which will raise the junction temperature to the limit. The curves illustrate the effect of fitting the tansistor on to a smeall heat sink.

## Circuit for Transistor Output Stage

The circuit of a Class A outpul stage is given in Fig. 4l.

Now can be seen an additional catuse of loss of supply voltage. This is the drop across the resistance in the emitter circtait. This resistance, and the transformer primary resistance, are generally arranged to drop around one volt, this being satisfactory from the point of view of stabilisation.

## Variation in Supply Voltage

What has one to do about the variation in volate that is bound to occur when a battery supply is used"? Dry batteries lose voltage whilst in use but in the interests of economy they lave to be used for a reasonable period of time. In the calse of a car radio the acemmator may drop below the nominal value and also it will rise under charge, and the radio is likely to be used under either circumstance.

The method of approach aga in depends on whether the maximum possible outpul from a given transistor is needed or whether a predetermined output from a conservalively rated transistor is aimed at. In the case of a receiver operating from a dry battery the former is likely to be the case. The procedur is then to calculate the collector current for the maximum permissible dissipation at the peak battery voltage and one can do no more than tolerate the fall in dissipation-and of power output-until compelled by the inadequacy of results to replace the battery:

The car radio case is likely to take the allemative line. Very likely the transistor will be capable of giving more power output than it is decided to use (after all one does not take more out of the electrical system than necessary), so the thing to do is to calcudat the collector current for the minimum output audio power likely to be ateeptable at the lowest voltage likely to be encountered. An output of two
watts is reasonable, and a 12 volt system is not likely to fall below II volts att times when the radio is in use. Then, allowing half a volt for curvature of the characteristic and one volt voltage drop in the series resistance : -

$$
\mathrm{Ic}-\frac{2}{10-.05}-.42 \mathrm{amps}
$$

and the base bias would have to be adjusted to give this value ol colleztor current. This indicates a dissipation ol $10 \cdot .42$, or 4.2 watts. The transistor chosen would have to be rated for a much higher dissipation than this, however, because the battery under charge whea the car is running is likely to reach

a voltage of It. The base bias will increase in proportion as it is drawn from the same supply and will catuse a simitar increase in collector current. So now

$$
\mathrm{l}=\frac{.42}{10}-\frac{13}{-.55 \mathrm{amps} .}
$$

and the dissipation $W=.55 \quad 13-7.15$ watts.
This has been warked out allowing for the drop of voltage in the emitter lead. but without taking into account the effect of the stabilising circuits on the base current. Wihhout such stabilisation it appears that a ransistor capable of a dissipation about four times the audio power required must be used for a range of voltage supply to be expected in a car radio. Allowing for normal stabilistation things are not quite so bad. but even then a permissible dissipation three times the audio power aceeptable as a minimum is desirable.

## Load Resistance

The load to be applied to the transistor can be estimated, assuming the ideal transistor chameteristics. which is near enough for the purpose. from the simple relationship of battery supply voltage and power output for which the circuit has been sed up. Thus,

$$
\mathrm{R}_{\mathrm{L}}-\frac{\mathrm{E}}{\mathrm{~N}}
$$

where $E$ is the battery supply voltage and Ic is collector current. Rut as above

$$
\mathrm{E}=\frac{\frac{1}{2} \text { Pout }}{\mathrm{E}}
$$

so by substitution

$$
\mathrm{RL}_{\mathrm{L}}-\frac{\mathrm{E},}{2} \frac{\mathrm{P} \text { out }}{}
$$

From this it is interesting to compute the circumstances under which the normal speaker impedances of 3 ohms and 15 ohms match the transistor directly without a transfomer. The 3 ohm speaker is not likely to be much use for domesti: or car radio with-
out transformer because at a power supply of 6 volts it matches at an output of 6 watts, and for 12 volts it is higher, actually 24 watts. The 15 ohm speaker appears to be more hopeful because on a 6 volt circuit it matches for an output of 1.2 watts and on a 12 volt battery the matching output is almost 5 watts.

## Thermal Runaway

It is with a power transistor that the dangers of thermal runaway are encountered. This phenomenon is due to the increase in temperature of the junction caused by the power dissipated. This increase in temperature itself increases the current flowing which in turn further increases the temperature. If the iunction cannot easily lose the heat generated this feedback effect can build up until the transistor destroys itself. Even if conditions prevent selfdestruction those conditions will cause distortion due to the resultant limitation of the current swing.

COMPONENT LIST FOR FIG. 41

```
R1-4.7 K!!
R2-1 K!? Dubilier (t Watt
R3-33!2 ,
C1-10\muF 6 v. Daly H2 5/1
C2-100\muF6%.Daly E2 49:1
TFR-Belclere BN 1827
```

Only in the small transistors is it wise to operate the transistor by itself. In other cases a heat sink is provided. This is effected by mounting the transistor in thermal contact (not necessarily in electrical contact) with a piece of metal and transistor manufacturers give guidance with regard to the amount of metal needed for various levels of dissipation.

The principle of design of the base circuit follows the lines of the audio circuits already described.


Fig. 42.-Collector characteristic-OC72.

Where large audio outputs are sought, however, it has to be remembered that a comparatively large power has to be fed to the base and in this case it may be necessary to design the driver stage as a power amplifier rather than a current amplifier.

## Design for an Output Stage

Now a low power Class A single-ended output stage suitable for addition to the receiver developed during this series of articles and following the circuit in Fig. 41 will be designed. This will not involve the complications of high power drive and the audio amplifier circuit as already described will drive it. This method of operation is not likely to be entirely satisfactory for permanent use because of its comparatively low output, but it is set up as a logical step in the process of design as it uses one of the two transistors to be used in the next step in Class B push-pull. In fact, if possible, it would be as well to buy a matched pair ready for the final design and use one experimentally in the present circuit before proceeding to the more complicated one. The transistor is actually a Mullard OC72 and it is operated-from the six-volt battery already provided for the rest of the circuit.

The permissible dissipation of the OC72 is seen from Fig. 40 to depend on the maximum ambient temperature under which it will be called to operate. For such cases as car radios using larger transistors and where the receiver has to work in confined spaces, this. factor has to be given great thought. In the present case, however, operating a well ventilated situation is assumed and a temperature not exceeding 35 deg. C. will permit a dissipation up to 100 mW . without heat sink. If the temperature is likely to exceed this the same design would be satisfactory except that the transistor would have to be mounted on a " sink." A loss of 1 volt in the emitter resistance and output transformer primary is assumed. The maximum possible output is needed, so

$$
\mathrm{Ic}=\frac{\mathrm{W}}{\mathrm{E}}=\frac{100}{5}=20 \mathrm{~mA}
$$

(To be continued)

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Many old receivers are in use in which the circuit is of th: "straight" type and which, therefore, do not lend themselves very easily to conversion for Band III reception. The best thing in such cases is to modify the video secticn making it conform to modern practice, and then to fit a two or three station tuner, or if preferred, one of the commercial turret tuners. How to do this is explained in the current issue of our ccmpanion paper PRACTICAL TELEVISION now on sale.

An improved Frame Sync circuit is described and will interest those who fecl they nould like better interlacing, whilst other articles in this issue deal with how to make a Skeleton Slot Aerial, Scanning and Synchronisation, a Mismatch Distributor, The Booster Diode, Tracing Obscure Faults and Servicing the KB LFT 50/60.
Also to be found in this issue are the usual features, Correspondence, Underncath the Dipole, News from the Trade, etc.

## The Audio Fair-1958

## detalls of some of this year's exhibits

THIS year's Audio Fair was more comprehensive than last year"s, and again depicted the many advances which have been mado in $\mathrm{Xi}-\mathrm{Fi}$ equipment. As usual the majority of apparatus on view consisted of amplifiers. pichups and associated gramophonc eqtipment. loud speakers and tape recorders. Many ofd friends were to be seen. as well as newconiers. and the various rooms which were reserved for demonstrations were, as usual. crowded to capacity. The following are some brief notes of some of the many items which were to be seen.

## Scotch Boy Tape

For the first time the entire range of the 3 M Company's "Scotch Boy" Magnetic Recor"ding Tapes was on show

The display of tapes featured the three "Scotch Boy" recording tapes on various spool sizes:

No. 111A Standard.-This has a standard brown oxide coating. on an acctate base. and is avaijable in spools of $200,600.850$. 1.200 ft . and other sizes for prolessional use.

No. 111 V Super PVC.-Designed for evtra strength and easy handling. With similar character-


The use of tas hoards is weil illustrated in this G.E.C. pre-amp.-
 the Fair. The Merven Mitas: Data Recorder mtilises Scotch Ro1. Tape. ant has a malti-chamnel sustem of recording.
because of the extra-thin backing. The " weather balanced" flexible superstrong polyester backing has outstanding mechanical properties over a wide temperature range and an extremely low rate of water absorption. It is available in spools of 300 . 900 . 1.275. 1.800 . 2.400 ft . and 3.600 ft .

Two new G.E.C. derelopments in high-quality boudspeaker technique-one a complete unit which will be marketed, and the other a design for home constructors-were shown.

A specially constructed $50 w$ amplifier and pre-amplifier-the " 88 -50"-is shown on this page. Offering improved performance and control range, the two units have inputs for a radio tuner. a magnetic or crystal gramophone pick-up. a microphone or a magnetic tape replay head. The pre-amplifier gives full playback lacilities from any known programme source, using simple and coonomic circuits. It is designed to give an output of 0.5 volts r.m.s. for maximum signal level. corresponding to the input required by the amplifier.

## Mullard

The Mullard stand featured the universally available World Series audio valves, together with a range of alloy junction transistors for batteryoperated audio amplifiers and pre-amplifiers.

The transistor range includes the OC16, an allmetal power transistor for the output stages of medium-power amplifiers; the OC72, which is available in matched pairs for push-pull output stages of low-power amplifiers; and the OC71 for use in pre-amplifier circuits.

## W.B. (Whiteley Electrical)

The complete range of Stentorian High Fidelity Loudspeakers from 8 in . to 18 in . in diameter, as well as special loudspeakers from $1 \frac{3}{4} \mathrm{in}$. upwards, fitted with the patented cambric cone providing a quality of reproduction not otherwise obtainable enedpt at many times the price. were shown.

Thi W.B. 12 Quality Amplifier is available with wither the "Standard" or "Major" Controls Units. embodying the most up to date designs and recently developed valves, and a special Whiteley output transformer. The Major Control Unit enables any pickup to be used with the W.B.12, and has a nine position selector switch, in addition to separate treble, bass, "Filter," "Slope" and volume controls.

A further important addition is the W.B. 6 combined amplifier and control unit, with 6 watts output, forming a compact unit size $11 \frac{3}{4} \mathrm{in}$. $\times$ $7 \mathrm{in} . \times 4 \mathrm{in}$. It has a selector switch for a range

## Further Notes on

REVIEWING the many readers' enquiries about - the Simple Geiger Counter described in the April issue, a few points have arisen which may be of interest.
First, the G-M tube G24 is obtainable from 20th Century Electronics Lid., King Henry's Drive, New Addington, Surrey, and the price is $35 /$-, postage extra. This lube measures 270 mm . in length, but type GI2 measuring only 145 mm . will work almost as well and can be neatly strapped to the side of the box. This lube also costs 35/-.

The type numbers of the batteries used COL230 and OL250) are those intended for deaf-aid sets, and suitable equivalents are Ever-Ready B106 for H.T., and either D18 or D19 for L.T.

The rectifier K840 is made by Sentercel, and the apparently over generous ratings are due to the fact inat the forward and reverse resistances were the primary considerations in the choice of this component. The valves CK 505 AX may be obtained from Service Trading Co., 9, Little Newport Street, Leicester Square, London, W.C.2, at $2 / 6$ each, postage extra. This firm also has most of the other components in stock.

Those having difficulty in obtaining the midget neon may be interested to learn that the majority of neon may be "ntine mains "circuit testers" contain nothing more than one of these neons with a series resistor and test prods attached.

A minor printing ergor in the wiring layout onlits the connection between the junction of Cl and R4 and the grids of $V 2$ and $V 3$. but reference to the circuit diagram will reveal the missing uire.
of inputs, full range bass and treble controls and separate volume control.

A new F.M. tuner was shown for the first time with a circuit of advanced design. An important feature is the use of printed circuit techniques to ensure maximum electrical stability, together with rigid mechanical structure, thus ensuring freedom from drift. The complete unit is housed in a shect steel case designed to match the amplificr.

## Collaro

Collaro L.td. presented their Conquest Record Changer. This revolutionary 4 -speed model is unique in that records of any dimension from 6 in . to 12 in . can be loaded and autamatically played. There is no finger to be reset and no protrusions which might cause damage to records. The Conquest also has provision for manual play at all speeds, and automatically switches off after playing the last record. A double action cam is incorporated in the change speed control, thus eliminating all possibility of damage to the idler wheel if the speed control is operated while the machine is running. The record changer spindle is designed so that there is a gentle lowering of records from the stack. ensuring hundreds of playings without damage to the centre holes. and a sensitive velocity trip mechanism operates on all modern and most old records.

Another interesting exhibit was the new improved Tape Transcriptor. Mark IV. which incorporates several modifications to the previous design.

## the Geiger Counter

## The Choke

A more detailed description of the home-wound ringing choke has been requested by some, but this is not critical with regard to resistance and number of turns. For example, the two prototype chokes were rapidly wound on an electric motor with 40 s.w.g. D.S.C. enamelled wire, and the resultant winding was loose and spongy in texture, whereas the Teletron product is neatly wound with enamelled wire which occupies only half the volume.

Any breaks in the wire occasioned during winding must be rejoined by soldering, and the joint well insulated by Sellotape or transformer paper. The end connections must be particularly well insulated from each other and the rest of the winding, or sparkingover may occur.

A reference was made to Government publications of use to those going prospecting for uranium on their holidays.

The following works can either be purchased from the "Geological Survey and Museum," Exhibition Road, South Kensington, London, S.W.7, or they can be seen in the library of the museum.
(1) Dines, H.G. "Metalliferous mining districts of South-west England " (Mem. Geol. Surv. G.B., H.M.S.O., 2 vols.).
(2) Davidson, C. F. " Uranium Deposits in Great Britain."
(3) Ponsford, D. R. A. "Radioactivity Studies of some British Scdinentary Rocks."

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# Is a Pre-war Set Worth Servicing? 

SOME DATA ON OLD MODEL RECEIVERS AND THE PROBLEMS INVOLVED IN SERVICING THEM

MOST service engineers have, in the course of their work. run up against this problem. I have had many of these sets brought to me for service. and it has been a difficult problem to decide whether to service the:n or to advise the owners to scrap them. In many of these cases. the reason for people wishing to have them serviced is because they like the cabinet. which has often. especially in the case of racliograms. been kept in a very good condition. Most of the pre-war cabinets were of good. sound timber and not the thin ply of which many present-day cabinets consist. Generally. when one informs the owners that the set is not worth the cost of the servicing and suggest they purchase a new one, they reply that having heard the new sets they cannot compare with their old one. This is mainly because they have become accustomed to the tone of their old set. which in the case of most of the pre-war sets. had a sharp topect in the amplifier and loudspeaker. thus giving what they consider was a mellow tone. When they hear the later day models giving a lot of top, to which their ears are not accustomed. they consider them harsh and not pleasant listening. My method of dealing with these sets is not to have any hard and fast rules about them, but to judge cach case on its own merits

## The Pros and Cons

In the case of people who I know cannot afford a new set, such as. for instance, old age pensioners, I do my best for them. if it is possible to put the set in order. with the chances of a reasonable further period of useful life. but I always inform them that owing to the age of the instrument I cannot guarantce that it will not break down again with in 12 months.

In the case of people who can afford to purchase a new sec. I point out that first!y the cost of servicing. including labour. new parts and ralves, may be quite high and perhaps be more than the set is worth. Also I tell them that as all the component parts have already had at long life and to replace the lot would be out of the question. there would always be a possibility of one of these components breaking down within a short space of time after the set was serviced. and therefore no guarantee could be given. If they do not accept these conditions then I turn the job down.

## Jobs to Reject

Should it be a case where they wish to retain the cabinet, then 1 suggest. if the set itself is in yery bad condition. that a new chassis be installed in the cabinet. Care should be taken herc to see that a chassis is chosen that does not necessitate cutting or drilling the cabinet for control spindle purposes.

Sets that are not worth while attempting to
service are those with the old type side contact valves. These valves are practically unprocurable and to try and replace with five- or seven-pin types means a considerable amount of time and trouble. The main reason is that the holes in the chassis for these side contact valveholders are considerably larger than for other types. This means plates being cut out for normal valveholders and fitted on the chassis.

Any sets using the old 4 volt heater type. especially the five-pin types, should be carefully considered. Some of these valves are impossible to obtain and to convert to a 6 volt range would not be an economic consideration.
Other types with the old large type band pass coils that have gone open circuit should be bypassed. Replacements for the coils are hopeless. Rewinding takes up too much time. and consequently the cost of servicing becomes excessive. Other sets that I suggest should be rejected are those with rubber-covered leads. where the rubber is either perished or in a bad state. To attempt to do this type of job means practically complete rewiring.

## Points to Watch in Servicing Old Sets

It is advisable in all cases to renew the electrolytic condensers. and also the rectifiers, valse or metal type. but it should be pointed out that then the H.T. voltage will probably be higher than before. This means that the bypass and coupling capacitors will be subject to a greater strain and as they are old there is a liability of their breaking down. When these electrolytics and rectifiers are changed. the set should be put on a soak test and checks made on all resistors for temperature rise. At any sign of this ozcurring, change the bypass condenser concerned, as it is probably leaking. which is often the prelude to short circuiting. It is also advisable to check voltages on valves. which will. in most cases. show up any other likely near failures. OId A.C./D.C. sets are those to view with great suspicion, as the heat radiated inside these sets is often much greater than in A.C. sets and this causes greater deterioration of the components. Mains dropping resistors or line cords should always be replaced. as this is where the greatest heat comes from.

## Remarks on Valve Changing

From the foregoing the reader can see that. unless it is being done as a personal favour. or if the owner does not mind the cost. it is not advisable to undertake these jobs. Even when the greates! care is taken and an extended soak test given. one cannot be sure that the repair will last for a reasonable time. without another breakdown.

It often happens with these sets, that a valve
has failed. and the correct replacement cannot be obtained, and a later type valve has been fitted with a base adaptor. Now it should be pointed out that. although the new valve is of the same type, has the same heater voltage and current, its characteristics are often considerably different. When fitting a valve of this description. always check up on anode. screen and cathode voltages. In the case of valves in tuned circuits. a slight alteration of trimmers may be necessary to allow for different inter electrode capacities of the old and new valves.

## International Car Radio Rally

TVHAT is believed to be the world's first International Car Radio. Rally will commence on June 25 th in Britain and 26th on the Continent. The main difference between this and other motoring rallies is the fact that participants will be given their instructions over the air from Radio Luxembourg. The prime qualification for entry, therefore, is that participants cars must be equipped with car radio. Instructions will be broadcast at frequent intervals throughout the rally in English, Dutch, French and German on two wavelengths-208 and 1293 metres.

The rally is organised by the Regional Motoring Club of South Holland and the Luxembourg Motoring Club. It is open to entrants from Great Britain, France, West Germany and Belgium, Holland and Luxembourg.

Starting points are Londoñ, Amsterdam, Paris. Brussels, Luxembourg and Bonn. It is planned that British entrants will spend the night of the 25 th at Dover, crossing the Channel the following morning. At Rheims they will join the main route, as will participants from Luxembourg and Paris. Those setting out from Amsterdam, Brussels and Bonn will have met at Eindhoven (Holland) and will join the main route at Dinant. A reception centre has been arranged at Chaudfontaine (Belgium) and the rally finishes at Luxembourg in the late afternoon of the 27th.

## l'assengers

In contrast with other international rallies, each car will be permitted to carry more than two passengers. The organisers stress that the rally is in no sense a test of speed and endurance and, while some simple skill and regularity tests will be included, they will not affect the nature of the rally as a tourist event involving moderate speeds and easy map-reading. Participants will. therefore, have plenty of opportunity of enjoying the splendid scenery through which they will pass. The 800 mile route covers some of the most beautiful parts of the Low Countries, the Ardennes. the Eiffel and Northern France.
On June 28th a winding-up party will be held and prizes will be presented. Among other attractions will be an entertainment provided by artistes of international standing.

## Entries

British motorists wishing to take part are invited to apply for further particulars to Competitions Department, Royal Automobile Club. Pall Mall. London, S.W.1. The latest date for entry is June 7th.

## Changing Loudspeakers

Many of the old sets had electrically energised speakers and replacements are out of the question. It therefore becomes necessary to fit a permanent magnet speaker. In this case an L.F. choke will be necessary, to act as H.T. smoother in place of the loudspeaker field coil. In fitting this make sure the inductance and also the current carrying capacity of the winding is sufficient. It is better to use as large a choke as space will permit than to use a small one that will overheat and perhaps burn out.

## New. Müllard Photo-cell

ANEW type' of photo-electric cell which will enable the cost of many industrial control and detection devices to be reduced substantially will shortly be available in quantity for the first time in this cóuntry.

The new cell incorporates a specially constructed photo-sensitive element of cadmium sulphide, and is remarkable for its extremely high sensitivity: it is, in fact, some 20.000 times more sensitive than the conventional p.ec. cell,

Even from weak sources of light it sill produce sufficient current to operate a large relay. direct without the need for intermediate amplification. The electronic amplifier circuitry assaciated with most devices based on conventional photo-cells is therefore made unnecessary, with consequent savings in equipment costs, simplification of design, greater reliability and casier maintenance.

Furthermore, the cell will produce the current necessary to operate a relay with only a very low applied voltage. This is made possible by a special form of construction in which the resistance of the cadmium sulphide element is effectively reduced by an interdigital pattern of copper strips.
The cadmium sulphide cells will be mounted on standard valve bases indentical with those of the present Mullard range of concentional cells. Thus, when existing equipments are modified, special head mountings or probes designed for specific installations can be retained.

## Advance Performance Details

Typical performance figures for a Mullard cadium sulphide cell with a photo-cathode of 1.8 square cm . effective area include the following:

From an illumination of 5 ft --lamberts, with a colour temperature of 1500 deg . K , the cell will produce approximately 20 mA of current for an applicd voltage of. 10 v . From the same illumination, but with a temperature of 2700 deg. K. the current is approximately 6 mA . Within the limits of permissible power dissipation doubling the applied voltage gives a four-fold increase in current. Maximum dissipation is 1 watt at 25 deg. C and 200 mW at 75 deg. C.

Dark current is extremely low: with 300 v . applied to the cell it is not greater than 2.5 microamps at 25 deg. C.

Spectral response range is 4500 deg . A to 8000 deg. A. covering the entire visible spectrum and extending into the infra-red, with maximum response in the yellow/red region.

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HHIS receiver was built primarily with the intention of providing a reasonably high quality reproduction of recorded music from both gramophone and tape. but a considerable amount of effort has been spent on providing a radiofrequency fecder to give results similar in quality to those obiainable from the gramophone. The feeder unit is the part of greatest intcrest in this series. as it sets out to overcome the disadvantages of Amplitude Modulation broadeasts on the Medium and Long waves.
quite a variety of answers to this. some of which are given below:

The principal disadvantage of F.M. is the limitations of the variety of programmes which may be received. The prototype of this series receives the L.ight Programme. the Home Service (any region). and Radio Luxembourg. but there is no reason for selecting these three except for personal preference. The coilcourse. the query is raised. why not use the new packs available give a wide range of choice of Frequency Modulation transmissions? There are packs avalable give a wide range of choice of
stations. and. by constructing the coilpack at


Fig. I(a).-Circull of the feeder section.
home. the constructor is not limited to three stations. In addition, of course, a variable funing system could be installed by suitable modification of the constructional details.

Another disadvantage of F.M. is that only local stations can be reccived with ease (and quality), and this makes difficulties for "fringe areas."

The reader should from the first be elear that this receiver is not intended to satisfy the most stringent demands of the High Fidelity enthusiast, but rather to provide a unit which, while being well above the normal standard of quality, Hoes not involve great expense or technical howledge to construct.

## The Circuit

The tuning of this receiver is carried out by an Osmor preset coilpack, of which there are several types available, to suit various requirements. When used with a conventional circuit, these coilpacks normally have adequate stability, but, as an added refinement, this receiver incorporares a stabilised H.T. supply for the oscillator section of the frequency changer. No previous stage is used, mainly because facilities are not available on the coilpack in use. but if it had been considered neccssary. this difficulty could have been surmounted. The circuit has been found in practice to have adequate selectivity and sensitivity. The frequency changer and


Underside. viaw of the second mait, which includes the section shown in Fig. I(b).
tuner sections are mounted in a separate comparment, sereened from the rest of the set. and it should be noted that. due to the screening, it is not possible to operate the set without a short aerial.

In order to provide an output signal of more than average consistency of amplitude. especially when recciving Radio Lusembourg. it was considered necessary to amplify the A.V.C. supply.

In order to provide an adequate range of amplification values to mahe full use of the amplified A.V.C.. two stages of variable-mu I.F. amplification have been used.


How the sections are housed in the upper part of the cabinet.

Following the detector stage the audio frequency signal is made available on a switch, in common with any other desired audio frequency signal. by which any one signal may be selected at will. A simple means is provided for regulating the amplitude of each signal. so as to allow them all to be adjusted to a similar level, to avoid the necessity of readjusting the volume control on changing from radio to gramophone, or vice sersa.

It was thought desirable to fit a simple scratch filter at this stage, and this is performed by a single R.C. filter network. The filter is continuously variable for the sake of convenience, and is arranged to attenuate all frequencies above a fixed cut-off frequency at a rate of approximately six $d b$ per octave. The cut-off frequency may be fixed at any value between $2 \mathrm{kc} / \mathrm{s}$ and $20 \mathrm{he} / \mathrm{s}$ by a knob on the front pancl. at the latter setting having no effect on the amplifier performance.

The signal is then passed to an unbiased cathode follower circuit. which enables the output of this unit to be connected to any one of the wide range of load values, as required.

In normal use. the signal is then passed down a coaxial lead to a Mullard 5-10; or similar high quality audio frequency amplifier. The output is by a Goodmans Axictte loudspeaker in a cabinet


The cor


View of the second witit.

edreceiver.

The Frequency Changer Stage
As has been stated earlier the tuning coilpack is an Osmor preset model. of which several versions are available the prototype using type l.LH. The circuitry of the coilpack has been included for easy reference, although the switching has been omitted for simplicity. The details of connections will be found in the instructions with the coilpack.

The valve, V6. is an Osram X.79, or equivalent. The hexode section is operated at an H.T. of 250 volts. decoupled by C20 and R31. An identical decoupling network is used on all stages. The H.T. supply to the anode of the triode is 150 volts stabilised: R37 is the anode load of the triode.

The aerial network. (18. R29, is intended to caclude $50 \mathrm{c} / \mathrm{s}$ signals from the aerial input. and thus climinate modulation hum. In some areas, where this type of interference is common. R29 should be replaced by a R.F. choke, as recommended in the coilpack instructions.

Note that A.V.C. is fed to this section of the circuit at half the level used for the other stages. This is brought about by the potential divider R34, R33. The network R32. C34. is inserted to prevent stray signals from the A.V.C. line causing interference in the grid circuit of V6. An identical network is used on all stages. C.19 is inserted to isolate the A.V.C. voltage from R32 from the carth on R29.

The oscillator uses a straightforward Colpitis circuit. this arrangement being perfectly normal in this application. Grid bias for this section is obtained in the usual manner for this arrange-


Linderside riou ff the section shown in Fig. $1(a)$. ment. C23. R30 prosiding the "leaky grid" circuit.

The anode signal output of the stage is fed sia 1.F.T. 1 to V7. The I.F. is $465 \mathrm{I}: \mathrm{c} / \mathrm{s}$, and the I.F.T. is a Denco type I.F.T.ll.

## The I.F. Amplifier Stage

The " earthy" end of I.F.T.I is fed with A.V.C., via the filter C25. R38. which is included
to eliminate stray signals and interaction. The other end is conmected to the grid of V7. and it should be noted that. by careful design. both this lead and the anode lead have been reduced to half an inch in length. Care should be taken when drilling the chassis to orientate the valveholder correctly so as to maintain this condition. The stage is fed by a decoupled H.T. of 200 volts. and is perfectly conventional in circuitry.

The valve is a Brimar 9D6 (Osram W77) or equivalent. This is a variable-mu R.F. pentode. on a B7C base.

The output of this circuit is fed via I.F.T. 2 to a similar stage acting as a second I.F. amplifier. The only difference in the circuit of V8 is that the A.V.C. supply signal is taken via a capacitor C30 from the anode. so no further comment is necessary.

## The A.V.C. Amplifier Stage

This circuit is similar to the I.F. amplifiers. but is fed by an R.C. circuit. R44, C30. This stage, of course. has no A.V.C.. so R44 is earthed. The vatve is an Osram Z 77 or cquivalent, although another W77 will work just as well without modification. The H.T is 200 volts, and the output is via I.F.T.4.

## The A.V.C. Detector Stage

The circuit of the A.V.C. detector may appear confusing at first sight. but is. in reality. perfectly conventional.

As is normal, the signal is fed to the anode of a diode whose cathode is held at a positive voltage. decoupled by C35. The A.V.C. signal is then taken directly from this anode. It will be recalled that the circuit commonly used is similar. except that the cathode of the diode is also used as the cathode of the first audio frequency amplifier. The present arrangement is necessary here as a higher value of voltage delay is required due to the amplification of the A.V.C. signal. This voltage is obtained from the

## VALIES

 potential divider R47. VR48, of which VR48 is used to control the voltage. - The procedure of setting this potentiometer will be deseribed later.

The value. V10a. is half a 6AL5 (Ostam D77 or other equivalent).

## The Signal Detector Stage

This stage is not quite the same as the circuit commonly used. because we take advantage of the fact that the cathode has only one use (instead of three. as is usual). and cmploy the diode in a simpler circuit. after the fashion of a half-wave power rectifier. The valve. V10b, is the other half of the double diode used as the A.V.C. detector. The choice of which section to use for which purpose is entirely a matter of convenience. but the screen between the sections should be earthed.

After detection, the audio signal is passed to an amplitude control. VR49. The reason for this is that it may be found that the signal obtained from the radio is considerably greater

than that from the gramophone in which case VR49 is used to preset the signal to a lower level in order to minimise the necessity for adjusting the volume control on switching from radio to gramophone and vice versa. This control is preset, and does not appear on the control pancl. It is very useful. especially in reducing the wear on the main volume control. and the advantages gained by its insertion easily justify the slight extra expense.

In order to ensure that no R.F. component remains in the signal after this stage, the control VR49 is decoupled on both sides by C36, C37.

## The "Gram-radio" Switch

When designing this receiver. it was thought that it would be useful to fit a number of audio frequency input sockets, from which signals could be selected at will by means of a switch,
but it is obviously superfluous to have any more than the two words "GRAM" and "RADIO" on the front panel. This will explain the apparent discrepancy between the diagrams. Further "Gram" channels are selected by turning the knob anti-clockwise. past the position marked.

It will be seen that each switch position has its own grid leak. There are two main reasons for this: first, to provide a load on the signal sources not in use, and to prevent the possible formation of static charges on the contacts, and secondly, to provide an internal arm of a potential divider. by means of which the gram or other external signals may conveniently be reduced in amplitude by the insertion of a series resistor. The reasons for requiring the latter are. of course, the same as those given for the insertion of VR49.
(To be continucd)

## LIST OF COMPONENTS

(The following items are required in addition to those Capacitors
needed for the Mullard Amplifier, Type 510.)

## Resistors

All are f watt minimum rating except where otherwise specified. Resistor Resistance Dissipation Function

| R29 | 10 k | $\frac{1}{2} \mathrm{~W}$ | Hunn freq. decoupler Dropper for V6/G2-3 |
| :---: | :---: | :---: | :---: |
| R30 | 33 k |  |  |
| R31 | 2.2 k |  | H.T. decoupler |
| R32 | 220 k |  | A.V.C. line blocker |
| R33 | 220 k |  | ) A.V.C. potl. |
| R34 | 220 k |  | $\}$ divider |
| R35 | 220 ohms |  | Cathede resistor, V6 |
| R36 | 47 k |  | Grid bias resistor, V6t |
| R37 | 33 k | $\frac{1}{2}$ W | Anode Load, Vgt |
| R38 | 220 k |  | A.V.C. line blocker |
| R39 | 2.2 k |  | H.T. decoupler |
| R40 | 220 ohms |  | Cathode resistor, V7 |
| R41 | 220 k |  | A.V.C. line blocker |
| R42 | 2.2 k |  | H.T. decoupler |
| R43 | 220 ohms |  | Cathode resistor, V8 |
| R44 | 220 k |  | Grid leak, V9 |
| R45 | 2.2 k |  | H.T. decoupler |
| R46 | 220 ohms |  | Cathode resistor, V9 |
| R47 | 150 k |  | H.T. potl. divider |
| \R48 | 50 k |  | H.T. potl divider |
| VR49 | 250 k |  | Volume control |
| R50 | 1 M |  | Grid leak, Vil |
| R51 | 220 k |  |  |
| R52 | 220 k |  | \} Grid leaks, V11 |
| R53 | 220 k |  |  |
| R54 | 150 k |  | Scratch filter network |
| R55 | 220 k |  | Cathode load, Vil |
| R56 | 1.5 k | 3 W |  |
| R57 | 1.5 k | 3 W | H.T. voltage |
| R 58 | 1.5 k | 3 W | droppers |
| R59 | 3.9 k | 2 W |  |
| R60 | 10 k | 2 W | $\begin{aligned} & \text { Safety resistor for V12 } \\ & \text { (see text) } \end{aligned}$ |
| R61 | 1 Mr |  | Discharge resistor |

## Coils

All I.F.T.s are Denco I.F.T.11, 465 kc 's.
Note : Connection instructions are included in the carton, and these should be closely adhered to. Pin 4 of I.F.T. 4 should be earthed.
Coilpack is Osmor "Switchpack" (any type) or home-constructed switch unit using standard coils. N.B.-No special instructions available from this magazine. Connection instructions in carton.

By H. C. Parr

EVERY owner of a tápe-recorder giving tape speeds of $3 \frac{1}{4}$ and $7 \frac{1}{2}$ in. per sec. must have tried the effect of recording music at the slower speed, and if he is accustomed to the wide frequency range of modern V.H.F. broadcasting and L.P. records. he must have come reluctantly to the conclusion that results are barely: satisfactory. The trouble is. of course. that the tape

output at this slow speed falls away rapidly at trequencies above about 1.000 cycles. and even with suitable compensation in the playback amplifier it is difficult to reproduce frequencies above about 5.000 cycles. The price one must pay for these frequencies is a doubling of the tape speed, and so of the running costs involved.

However. under certain circumstances to be mentioned later. it is possible to emphasise the upper frequencies while recording at $3 \frac{3}{3}$ in. so as to make up for the loss of output compared with that at $7 \frac{1}{2}$ in. and so obtain results identical with those normally achieved at the faster speed. Fig. 1 shows the circuit of a very simple device which does this. It can be built up in a fey minuics on the type of socket pair which is used for the "extension speaker" or "gram" connection of a radio.

## The Apparatus

Fig. 2 shows the appearance when completed.


Fig. 2.-Layout of the parts.
If desired a small wooden box can be made with one side open so that the sockets can be screwed on to it with the "works" inside. and a small hole for the leads to be taken through.

The appearance will then be as in Fig. 3.
The reader must suspect by now that there is some snag in expecting so simple a device to achieve so much. and he is not mistaken. The drawback is a reduction of 20 db in the available input. and so in general the device is useless when recording from a microphone. for the


Fig. 3.-The finished unit.
resulting signal will not $b=$ strong enough to load the recorder fully. This does not apply when recording from the radio. whether the output be taken from the extension speaker terminals or direct from the tuner. . The plugs shown in Figs. 2 and 3 are connected to the appropriate sochet on the radio. and the output from the sockets shown is taken to the rccorder through screened cable. If sufficient signal is not available when it is applied to the "radio" input. then the "microphone" input can be used.

## Precautions

There are two precautions to be observed in using this device. Firstly, the terminals 2 and 4 must be connected respectively to the carthy side of the radio and recorder. or hum may be introduced. Secondly. since the purpose is to make the tape response at $3 \frac{3}{3} \mathrm{in}$. equal to that at $7 \frac{1}{2}$ in.. the playback amplifier. if it provides compensation for both these speeds. should be switched to the " $7 \frac{1}{2}$ " setting. If the switching is linked to the speed change mechanism some modification will be needed for the amplifier to be in the " $7 \frac{1}{2}$ " setting while the tape runs at $3 \frac{3}{3} \mathrm{in}$.
There will be no, great difference in the behaviour of the recording level meter or magic eye. for in normal music the energy level is quite low at the frequencies involved.

The results at the slow speed. provided the recorder does not suffer from wow or flutter troubles. are then quite indistinguishable from those obtainable at $7 \frac{1}{2} \mathrm{in}$. If the machine is used frequently for making permanent recordings from the radio. a considerable saving in tape costs will result.

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| 1N： | 9 | $61^{*} 1286$ | T4i $12 i_{1}$ | $21^{-}$ | 61 BT 15 － | lapal | 89 | 1：4＂x4 | 10 | E＇VR1 8 | KT\％ | 128 | PLCit\％ 136 | ［20 | $2 ;$ | IFte | 17. |
| 1R： | $8 \cdot 9$ | （iFl： 186 | 7rt 761 | 90\％ 146 | G18120 $15-$ | 114146 |  | 2ccu | 108 | E\％ 50 | KTWfil | 9 | PL： 8 U36 | ［－4 | 27 | ＇F゙N0 | $10-$ |
| 14， | 8.9 | fild $15-$ | R19： 86 | 2H1： $23-$ |  | LWF：3 | 99 | Ec¢ |  | H＊ 146 | Ктい6． | 19 |  | 1－4 |  | 1 PR | 10 |
| ＋ | $8 \cdot$ | 6FI． 156 | ！1811 i；14／9 | －11＊ 2361 |  | 1）191 | 88 | E（F×） | 126 | $\mathrm{Fl} 1: \% 146$ | KTWb： | －9 | PLW $\quad 96$ | $1 \div 1$ |  | 1Fs： | $10-$ |
| \％ | 10.6 | 646 3－ | 1n＇1 18 － | －nt 236 | ：30： 88 | HFW | ${ }^{6} 6$ | 131 | 126 | N（1：H6 148 | K＇TZ41 | 9 | Plas： 11 | 1－： | 23 | 1.41 | 6 |
| dis | 12.8 | 6JられT 66 | 1410 | －21p： $21-$ | ：314 86 | W1：3 | 85 | Eu＇H | 21 － | 1）${ }^{\text {a }}$ | КТ\％＊＊ | $10-$ | P1，N－ 21 － | l： | 23 | 1．－11； | 9－ |
| ：34 | 108 | 6J6 76 | lırl 23 B | －014 236 | $\because 6.5$ | DF9\％ | 96 | Pt ${ }^{\text {chat }}$ | 96 | 126 | 1．64 | 68 | PM：A 10 | 1.8 |  | 1184 |  |
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| ant | $8 / 6$ | GだNはT 1218 | 1いL। 11515 | 3．1．5：T 9－1 | AC＊PEX | $1 \mathrm{H}^{\text {H7 }}$ | 76 | ECLhat | 126 | （17\％－ 116 | MH＋ | 88 | 19：4M196 | 174 |  | ＇s | 236 |
| F64： | 818 | おKㅎ．s 186 | 1619， 81 － | \％y 106 | 21 | $1)+1107$ | 106 | $1 \times 1 \times 0$ | 136 | H：ar 49 | 31114 | 186 | PMobrz 156 | C＇1析 | 13 ＇ | C1：1 | 15 |
| Y：30＇t | $8 \cdot 6$ | fial 15 － | lorlt 18 ¢ |  | At ${ }^{\text {a }}$［1id | 11171！ | $9{ }_{8}{ }^{\text {a }}$ | L． $\mathrm{F}^{\text {ct }}$ | 21. | H6ii $10-$ | M11速 | 126 | ア「ゴ3188 | $115:$ |  | 1 F |  |
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ASTANDARD method of comparing and checking frequencies in the laboratory is by means of Lissajous Figures on the screen of a cathode ray oscilloscope. For the experimenter. who does not possess an oscilloscope. or who has only onc which is ticd up in another part of the rig at the moment when he wishes to check on some frequencies. an alternative and simpler method is desirable. In this connection it is ironic to note that in the professional laboratory, where frequency standards are expensive and reliable. there is likely to be a sufficient number of C.R.O.s for monitoring their accuracy; but that the amateur. whọse oscillators and signal generators are likely to be home-made and distinctly substandard will seldom have more than one C.R.O.. and quite often not even that. It is thought. therefore, that the instrument described in this article will be of great use both to the serviceman and to the home constrictor.

In the first place. any such device, to be of value over a wide range of frequencies, ' must make use of the cathode ray tube in some form or another. No other device can offer the speed and freedom from inertia necessary for H.F. use. But fortunately the cathode ray tube is available in a cheap and simple form perfectly suited to frequency measüring systems. a form requiring no EHT. no time-bases, no focusing devices. and no deflection arrangements. I refer to the "Magic Eye."

## The Principle

Fig. 1 shows the principle of the device. The tube here is a 6 E 5 . One frequency. f 1 is applied to the triode grid. and appears on the target grid via the amplifying section of the magic eye. The other frequency $f 2$ is applied to the target anode. If the two frequencies are approximately equal. this will result in periods- when the target grid and anode are in-phase, alternating with periods when they are out of phase. The rate of alternation will be the beat frequency. the difference between $f 1$ and $f 2$, and this will. of course, be clearly visible in the form of a regular flickering of the display. If we adjust the variable frequeney to obtain zero beat. it is then equal to the comparison frequency.

This is simple enough. and quite useful. But the great usefulness of the principle in practice depends upon two further facts. The device works not only when $\mathrm{f} 1=\mathrm{f} 2$. but also when $\mathrm{fl}=\mathrm{k}: 2$ ( or when $\mathrm{kfl}=\mathrm{f} 2$ ). where k is any whole number from one up to ten or twelve: and every amateur has at least one good

> A Simple Method of Frequency Checking Which Does Not Call for the Use of a Cathode-ray Oscilloscope. The Arrangement Described Here Makes Use of the Standard "Magic Eye" Tuning. Indicator By A. M. St. Clair
frequency source. the $50 \mathrm{c} / \mathrm{s}$ mains. Hence, with the aid of a practical form of this instrument. we can use the mains to determine the accuracy of an audio oscillator in steps of 50 cycles up to $500 \mathrm{c} / \mathrm{s}$; and hence. by means of a simple auxiliary oscillator. we can proceed up to $5.000 \mathrm{c} / \mathrm{s}, 50,000 \mathrm{c} / \mathrm{s}$. etc. Likewise. from a single crystal oscillator, we can work downwards. in small enough steps to obtain excellent calibration, in checking a signal generator.

## The Circuit

A suitable cireuit. using a 6 E 5 . is shown in Fig. 2. Here a triode is added to provide amplification for the f2 signal. Any valve you have available ir the spares' box will do, and there, is no need to work under linear conditions. A pentode may be used, either triode-strapped or otherwise. Far from needing linear amplificts. this system works better when the distortion is high. particularly when beating against harmonics. It is for this reason that rectifiers are wired into the grid circuits. They are for the introduction of deliberate distortion in order to sensitise the apparatus to harmonics.

## An Alternative

An alternative scheme is shown in Fig. 3. 2 wing an EM34 magic eye. In this version, the signals are mixed in the common anode load, and the beat is fed to the eye through a low-pass filter. This is advantageous where R.F. signals are being monitored. since no R.F. reaches the eye itself. The possible frequency range of this variation is limited only by the valve used in the earlier stage (mixer). and the style of wiring. Since this particular form of instrument is sensitive to ripple

and variations in the H.T. line, a neon stabiliser is included. This limits the H.T. on the mixer to the lit voltage of the tube used; this is preferably a 120 volt tube. but even an 85 volt one will serve. The value of RI should be adjusted experimentally to a point where the neon is just stably lit. and this resistor should be heavy enough to carry the full neon current as well ds
point. Advance the oscillator dial to a nominal $100 \mathrm{c} / \mathrm{s}$. and repeat the process. Working in this way you will dbtain an accurate point on the dial for every 50 cycles up to $500 \mathrm{c} / \mathrm{s}$. Beyond that. another oscillator is required: but it does not have to be accurate or even calibrated. Of course. if you have a second oscillator also requiring calibration. it will serve and each oscillator will help in the calibration of the other. But if you have to build the auxiliary oscillator keep it simple-a one-valve transformer coupled or R.C. jobbecause it does not even liave to be stable for more than a tew seconds on end.

The set-up is as in Fig. 4. The auxiliary oscillator is wired across the f? terminals. in place of the oscillator under test. The latter is wired so that it may be switched to the fI terminals alternatively with the A.C. source. The ausiliary is set to $500 \mathrm{c} / \mathrm{s}$ by beating with either the mains or with the oscillator under test. which is now accurate at $500 \mathrm{c} / \mathrm{s}$. With the switch in position 1. the oscillator under test is made to beat with consecutive harmonics of the auxiliary. giving accurate calibration points at 500.1 .000 . $1.500 .2 .000 \mathrm{c} / \mathrm{s}$. and so on up to 5 or $6 \mathrm{Kc} / \mathrm{s}$. If a finer division is required. the atuxiliary oscillator will have to be set to a lower frequency. say: 100 or $250 \mathrm{c} / \mathrm{s}$. To cover a wider range. the auxiliary should be modified to work at a higher frequency. Before each reading. immediately before. the switch should be set to position 2. and the auxiliary thus checked against the mains.

The proccedings for checking. either up or down. against a erystal standard. are exactly the same in principle. The comparator finds a very useful application. rather less laborious. in checking an oscillator against a known standard. Here. we merely beat the setting of the uncalibrated oscillator against the setting of the standard without using either harmonics or auxiliaries. Broadcast stations can perform useful scrvice as " known standards.
the valve supply. Otherwise, the circuit. for the same reasons as given above-no linearity requirement-is singularly uncritical. If a stabilised power supply is available there is no need to include the neon stabiliser. in which ease the double triode may be given the full soures voltage, with an attendant increase in sensitivit.

## Using the Unit

A word or two on how to use the equipment. Let us suppose that you have a homc-built audio oscillator. such as has been often described in these pages. and that you wish to make an accurate calibration of its dial. Apply a voltage of about 2 to 5 volts A.C. derived from the $50 \mathrm{c} / \mathrm{s}$ mains via a suitable transformer to the fl terminals. Apply the output of the oscillator to the f 2 terminais. Adjust the output of the oscillator to a suitable level, and set it to a nominal $50 \mathrm{c} / \mathrm{s}$ as shown on the dial. A beat will be seen on the eve. Adjust the oscillator for zero beat. and mark the dial " 50 e/s" at that


Fig. 1.-This circoil illistrates the principls of the arrangement describad here.


Fig. 3.- A circuit for an alternative scheme.

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#  <br> MODULATING WIITH THE CLAMP TUBE <br> By O. J. Russell; B.Sc.(Hons.), G3BHJ 

ONE of the reasons why the clamp tube is a permanent feature of the present-day transmitters is the fact that it may readily be used to provide modulation. This is in addition 10 its value and original purpose of providing a "safety" device that protects the P.A. valve from excessive dissipation when drive or bias fails.

To elucidate the mysteries of "series sereen." "gating" clamp and similar modulators. it must be realised that they are all forms of screen modulation. Consider the simple transformer coupled sereen modulator of Fig. 1. This may be simplified to the choke coupled modulator shown in Fig. 2. If the chole is replaced by a resistor, we get the basic "clamp tube " modulator of Fig. 3. Thus the clamp tube system is purely "resistance coupled" scrcen modulation. Viewed in this way, there would ha:dly sem to be any need to stress this. However. a Iegend seems to. have arisen in which "clamp medulation" is regarded as having esoteric and bonderful properties quite unrelated to screen modulation, and presumably conferring mysterious and nebulous benefits not to mention greater efficiency, than ordinary "old fashioned "sereen modulators.

The cautious reader may now in fact begin to suspect that " simplifying " screen modulation down to a resistance coupled version may mean a sacrifice of efiiciency. This is in fact the case. and a simple clamp tube modulator is not as effective as a straightforward transformer coupled.


Fig. 1.—. 4 comemional transformer coupled weren modutation circuit.
screen modulator as shown in Fig. 1. Before considering this further. it is as well to consider the limitations and properties of sereen modulation from first principles. Without repeating the "step by step" detailed arguments which the writer has previously given. the following points should be noted. Seicen modulation is an "efficiency" method. and the simple forms suffer from the following limitation . . . that the actual R.F. carricr output is one-quarter of the carrier output that the P.A. valse is capable of giving under normal C.W. conditions at the same H.T. voltage. Due to efficiency considerations the maximum carricr output the P.A. valve is capable of giving at " all out " conditions is onc-half that obtainable under C.W. conditions. To make this clear. if we have a P.A. stage running flat out at liull anode dissipation at 100 watts under normal C.W. conditions, then with the same H.T. voltage under screen modulation conditions we shall only be able to run at 50 watts input, and at half the R.F. efficiency of nomal operation. Hence our actual R.F. carrier output is that of a normal 25 watts P.A. stage. If ve run under maximum possible dissipation and output, a substantial increase in H.T. voltage is necessary. Thes to takc a familiar example. the humble 807 needs voltages of some 1.200 volts for " good "cficiency modulation. Below some 1.000 volts of II.T. the characteristics of the tube are not fully exploited. This is a far cry from the "simple low power rigs" that utilise " simple and economical " sereen or clamp modulators. Thus some of the 10 watt topband rigs with clamp modulators are equiva-


Fig. 2.-- 1 choher roupled sureil mochilator.
lent to, say. 5 watt anode modulated rigs. Thus a 6 V 6 " clamp tube " might just as well be utilised as an anode modulator when it would be capable of directly anode modulating at least a 9 watt P.A. stage running at full C.W. efficiency.

## Advantages

Considerations such as the above thus tend to deflate the much boosted value of " clamp tube" modulators for users of simple low power rigs Thus the Tx running at 40 watl P.A. input under clamp modulation. is at best equivalent to a 20 watt P.A. stage. and if a 6 L. 6 c!amp tube is


Fig. 3.-The conventional clamp ture modulator is. in effect, a resistance coupled screesi modulator.
used that could modulate a 20 wati P.A. stage. What benefit is clamp tube modulation therefore. and surely it gises "better" results than that. At the risk of an explosion from the "defenders of clamp tubes." the efficiency figure for simple clamp tube modulators are as given. The chief advantage of the clamp tube modulator is that it is a simple system giving good speech with very simple audio equipment. To counterbalance the loss of carrier output. no separate modulator, powerpack and modulation transformer are required. Thus screen modulation systems merely require a small speech amplifier ahead of the clamp tube to provide clamp tube modulation. However. the fact that a simple system will provide good speech does not mean that a price is not paid in the form of loss of carrier. Moreover. the "efliciency" arguments indicate that the clamp tube system is hardly worthwhile for the low-power man. whereas a high-power C.W. rig can still run a respectable, though reduced. carrier output. An input of 150 watts clamp modulated. for example. equals a 75 watt anode modulated input . . at least ideally. Thus. with high inputs. a clamp tube modulated rig can give a respectable amount of modulated carrier. Note. however. that there is a hidden fallacy to beware of. Just because one can run 150 watts of C.W. that does not mean that one can run 150 watts input of clamp tube modulated telephony. To be able to do that. the P.A. valve would have to be "oversized," and capable of some 300 watts of C.W.
input. Thus the owner of a pair of 807's capable of just 150 watts of C.W. input would be able to run only about 75 watts of clamp modulated P.A. input. An amateur using an 813 in his P.A. stage. and with a power pack capable of some 2.000 volts output. could readily run an input of around 150 watts of clamp tube modulated input.

The above may seem to be discouraging to the amateur who has hoped that clamp tube modulation can automatically be applied to provide full modulation of his normal C.W. input. Unless his normal C.W. input is an unduly conservative oric. both efficiency and power input must be reduced to obtain effective modulation. With this limitation. effective speech may be readily obtained. This leads us to consider how the amateur may "set up" his clamp tube modulator system for good modulation.

## Satisfactory Circuits

As indicated. the atdio input from a small speech amplifier (Fig. 4) is applied to the grid of the clamp tube. The circuit of Fig. 4 should be adequate to modulate the usual clamp tube circuits. It is best to adjust such systems by running the transmitter into some form of artificial load. According to the power level to be expected. a 12 watt car bulb or a 25 , or 50 watt domestic light bulb should be tapped in across the aerial tuner in place of an aerial. Load up the full C.W. input with the clamp potentiometer turned for full carrier input. i.e.. with the clamp tube fully cut off. This is the normal C.W. operation, and the electric lamp used as artificial load should be lit to full brilliance. With the receiver aerial removed. and the gain turned down. it should be possible to monitor the carricr. Alternatively a pair of phones and a germanium rectifier plus an R.F. pickup coil should enable the speech to be monitored. To adjust to approximately the correct clamp tube setting for modulation, turn the clamp tube bias potentiometer so that the clamp tube starts to conduct. This will be shown by the P.A. plate millianmeter falling back from its normal C.W. reading. Ruthlessly reduce the P.A. plate meter input to half the normal C.W. current reading. This is a good position to start modulation tests. To attempt modulation. cautiously advance the audio gain of the speech amplificr feeding the clamp tube. and listen to the quality of the speech on the monitor. If good. clear speech is obtained. observe the lamp used as a load. This should flicker upwards in brilliancy on speech. If no noticcable flicker is obtained. the moduation percentage is quite low. If this low modulation percentage gives very loud headphone signals on your monitor. reduce the gain or pickup of the monitor so that speech is reproduced at a low level. To avoid confusion an assistant may be pressed into service to speak into the mocrophone. or even a broadcast programme may be utilised as the audio input. provided one is sure no radiation of R.F. from your transmitter is occurring. Finally, one should be able to get enough audio gain to produce noticeable llickering of the load lamp. However. watch the P.A. plate meter.
(Continued on pag? 305)
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It will be found that with enough audio input. and with adjustment of the clamp tube bias control the P.A. plate meter will kick strongly. This is an indication of incorrect bias setting. and not (as some imagine) a demonstration of "modulation." This warning is necessary, as some operators with a kicking meter gleefully quote " peak modulated input " readings from the " maximum kick height" and delude themselves into believing they are running very efficiently. Actually the kicks are a warning that adjustments are not right. If the meter kicks violently upward, adjust the clamp tube bias potentiometer so. that the P.A. draws more current. If the P.A. stage is drawing too much current. the P.A. meter will kick downwards, and the bias control for the clamp tube should be adjusted so that the P.A. stage draws less current. Finally, with the P.A. stage adjusted for minimum kicking, if the audio input is increased too much. kicking will take place irrespective of the bias control setting. In general the P.A. meter kicking slightly upwards seems to be a favourite position for good modulation and efficiency. However, this position is not far from the original trial position obtained by turning on the clamp tube so that the P.A. input is around half the normal full C.W. input. Cautious adjustment on the above lines. coupled with monitoring the signal on a simple crystal monitor will reveal much more about clamp modulation than might be suspected. If a cathode-ray lube modulation monitor is available. the results will amaze those who imagine that a wildly kicking P.A. meter is a sign of a high modulation depth, as it will reveal a very lopsided modulation envelope. Moreover a cathoderay modulation check will reveal that the modulation depth falls short of the 100 per cent. mark, no matter how much distortion is tolerated.

The fact that the modulation depth does not


Fig. 5.-Transformer coupling combined with the conventional clamp valve permits of more effective screen modulation:
exceed the 100 per cent. marh. or in fact with the simple clamp tube modulator falls far short of it, is even hailed as an advantage by some. The severe clipping that occurs with excessive audio input produces exactly the same pernicious effects as overmodulation. viz., splatter and spread. The clamp tube modulator merely provides these effects below the full modulation mark. in at any rate the simple circuits generally described. As these effects are clearly audible to


Fig. 4.- A simple two-stage speech amplifier preceeding the clamp valve is adeguate to give good modulation from a chystal microphone. The equrrent taken need only be a few millianps at, say, 250 volts. The preamplifier mal senerally be fed from the transthitter power supplies withour need for? a separate power pack.
anyone with a monitor and a clamp tube modulated transmitter. the "advantage" of " not overmodulating " is largely illusory

What is serious is the fact that one does not obtain full modulation with the simple clamp tube circuit. Even if the anode of the clamp tube could provide distortionless audio even when its anode swung down to zero volts. the carricr would still not be 100 per cent. modulated in-a downward direction. This is because the screth must be suung appreciably negative to reduce carrier output to a full stop. With the generally used clamp tubes. an inspection of their triode characteristics reveals that generally the limit of reasonably distortionless swing in the negative direction is at point where the screen and clamp anode are at a potential of around 50 vols positive. This seriously limits modulation depth ideally to some 66 per cent. in the downward direction for linear modulation. It is in fact why the "best" position of clamp bias adjustment is set so that there is a gentle upward kicking of the P.A. meter. By this means the upward or positive modulation is at its full level, with slight clipping of the negative peaks that would otherwise swing below the 50 voll mark. This somewhat increases distortion. but gives a somewhat bigher effective modulation level

## Full Modulation

Once adjustments reach the "somewhat " class. they are obviously at the mercy of the individual temperament of the operator. and tend to lead to the wildly kicking plate meter of the man who thinks he is getting his money's worth of modulation if the meter swings violently. This poses the question of whether there is a clamp tube circuit that will modulate fully. This question is settled by combining classical transformer coupled modulation with the clamp tube principle as shown in Fig. 5. This circuit will behave precisely as a normal clamp tube controlled P.A. By using a small audio coupling transformer. say: a step-up ratio of around two to one. the screen may be supplied with ample power. Where a


Fig. 6.-A small b.-passed auxiliarr dropper resistor (Rd) enables filler modulation to be obtained. If $R d$ exceeds 1,000 ohms. the screen resistor shorld be redured in salne to compensate.
small amount extra of modulation depth is still. needed, an auxiliary by-passed dropper resistance may be inserted in series with the screen feed. as shown in Fig. 6. This by-passed series resistor cannot be made too big. as it will result in an excessive drop of screen potential unless adjustments are made to the main screen dropping resistor value. If the resulting increase in " standby ${ }^{*}$ current may be tolerated. it is quite practicable to use an ausiliary screen dropper resistor by-passed for audio. and obtain full screen modulation with purely resistance coupled clamp tube modulation. This is illustrated in Fig. 7. and is recommended as a simple improvement that may be readily applied to existing " simple" clamp tube modulator circuits.


Tig. 7. - Abr-passed auxiliary screen dropping resistor will enable evell the simple clamp tube modulator to give deeper modulation. The clamp tube operates at a higher voltage than the P.A. screen, and is thus enabled to swing the screen wolts to nearly zero volts. The auxiliary resistor should be a few thousand ohms.

## News from the Clubs

I:ALING; YOUTH RADIO SOCIETY
THIS club, whose main interests are : (1) Radio Construction. 1 (2) Radio and TV Maintenance. (3) Amateur Radio Communication. would welcome new members. The meetings are held every Monday, $7.30 \mathrm{p} . \mathrm{m}$. at the Brentside Centre fold Nursery Schooll. Bordars Road. Hanwell. W.7. The club is sponsored by the Ealing Youth Sub-committee. and is open to boys between the ages of 14 and 20 years.

## BRIGHTON AND DISTRICT RADIO CLUB

Hon. Sec.: R. Purdy. 37. Bond Street, Brighton, I.
MORSE classes and "Fundamentals" classes are held on all M evenings where possible to fit in with the exisling programme.

As always, visitors and prospective members are invited to

- drop in." May fixtures are as follous:

Tuesday. May 6th-Multard Sound Fitm on the manufacture of television cathode ray tubes entitled " Made for Life."
Committee meeting to arrange future programme.
Tuesday. May l3th-it talk by Mr. H. R. Henly, title to be announced.
Tuesday, May 20th-Mr. J. P. Clement, the " C.R.O."
Tuesday, May 27th-National Field Day. Final detaits to be arranged.

NORTHAMPTON SHORT WAVE RADIO CLUB (G3GWB) Hon. Sec. : S. F. Berridge (G3!TW). 20. Ethel Street. Northampton.

$0^{\prime}$WING to altered domestic arrangements of the owner of the premises. meetings are still held at the Club Rooms. Allen"s Pram Works. 8. Duke Street. Northampton, but on Thurwiarinstead of on Fridays. It is hoped to revert to the usual Friday meetings at the end of October. A No. 19 set is being prepared for portable operation during the summer months as "field days " are planned using the Club call-sign G3GWB P.

TORBAY AMATEUR RADIO SOCIETY
Hon. Sec.: G. A. Western (G3LFL). 118. Salis'sury Avenue. Barton. Torquas:
$T$ HE Annual Dinner and Social held at the Oswalds Hotel. Torquay'. was attended by over 50 members. XYL's and visiting hams and proved a marked success. The President. W. B. Sydenham (G5SY), expressed his pleasure in having such a well attended function. this being responded to by the Acting Chairman. F. D. Cawley (G2GM), who in his brief review of Club activities commented favourably upon the large increase of members during the past vear. He extended a warm welcome to the committee members of the Eveter R.S.G.B. Group. whic:t included the county representative. B. Muaro (G3FLK).


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By Gordof J. King, A.M.I.P.R.E.

AS an introduction to the commerial application of transitors. We deal this month with the Cossor Transistor Record Player (lig. 1). While this instrument serves only as a record player and has no provision for the reception of radio signals, it represents admirably the wa! in which transistors are now used for the amplification of audio-frequency signats.

The full circuit is given in Fig, 2. trom which readers will have little difficulty in following the signal path, in spite of the less familiar symbols of the transistors. From recent articles on the transistor, it will be recalled that. like a thermionic valve, a transistor sequires a D.C. bias supply. and that the bias has to be arranged in the reverse direction between the collector and base.

In this sense, the transistor can be looked upon as two crystal diodes formed between the emitter and base and the collector and hase (in the circuit at Fig. I the letters B. C and E around the transistor symbols represent base, collector
and emitter respecticly). These three points ate often likened to $t=$ electrodes of a triode vale as follows: colle for and anode, base and grid and emitter and cathode.

The forward dirftion of current in the emilter: hase circuit due to the biasing gives this cireuil a lower resistance, while a relatively higher resistance is possessed by the collector/base circuit which is biased in the reverse dircetion. When the current in the base/emitter circait increases. the current in the base/collecinr circuit also increases, but to a greater evicont. Gienerally speaking, this action is promoted by the emission of so-called positive holes from ine emitter to the collector circuit with a consegrem lowering of the resistance of the base/collecond circuit.

The signal is applied to the transistor, so ats 10) Cause variation of the negative current in the base/emitter circuit. which results in an equal or greater variation of current in the base

lig. 2. Theoretical circuir of the revorl pleacer.
collector circuit. Power gain is thus secured by reason of the applied signal causing a current change in a low resistance circuit white promoting a similar current change in a high resistithte circuit. With these brief facts in mind we can investigate the Cossor circuit.

## Circuit Description

The signal from the crystal pick-up is coupied to the base of the first transistor (VI) by way of the volume control RI (note that in this case the slider of the control is connected to the pick-up and not to VI as it would be with a thermionic valve), resistor R2 and capacitor ( 1 . These serve as equalising and matching conponents from the pick-up to the transistor. The signal is thus applied to the base. which. as


Fig. 3.-The single-conded push-pull output stage, using a rapped H.T. battery:
already mentioned may be likened to the grid of a thermionic valve.
From the signal point of viex, the collector of $V I$ is at chassis potential (via (3). and the output signal is taken from the emitter circuit. From the valve aspect, this means that anode is at chassis potential from the signal point of viex: and that the signal is taken from the cathode. In other words. the stage is arranged in form analogous to the cathode-follower, and is usually referred to as a grounded-collector amplifier.

The advantage of this arrangement lies in the fact that it has a relatively high input impedance and can be matched into a pick-up circuit, for example : without calling for a transformer. Its output impedance is low houever but this is often desirable as a matching attifice when the inpul impedance of a proceeding stage is tairly low. such as the input impedance of stage V2 in the Cossor circuit.

The second stage in fact. consists of a grounded-emitter amplifier. Here the inpait impedance at the base is lower than the output impedance at the collector. and to avoid low frequency atentation as the result of the lou impedances involved coupling between V '1 and V 2 is made by way of an $8 \mu \mathrm{~F}$ capacitor C2. It will be appreciated of course that this
slage has a distinct resemblance to a signal triode stage in which the cathode is grounded and the signal is taken from the anode.

The current variations in the collector/base circuit and the power produced by them is abstracted by the driver transformer the primary winding being connected in the collector circuit of V. The transformer is designed to drive the push-pull class "B " output transistors V3 and V4 to provide an atidio output of some 300 mW with very little distortion.

## The Output Stage

The output stage is worth investigation since it represents at fairly recent development (in) valve circuits as well as transistors) which dispenses with the usiual split-primary transformer in push-pull output stages. It is described by the curiously coniflietingiferm" single-ended pusinpull output.:
Let us first looh ate idea in terms of valves with which we are possibly more familiar. In Fig. 3 are shown the two output valves VI and Vf. but instead of being connected across the primary of a centre-tapped output transformer. they are effectively connected in series with the H.T. supply. The loudspeaker is connected between the anode/cathode junction and a centre soltage tap on the H.T. supply. On the face of it. it may appear that the loudspeaker is going to suffer a burn-out. but if the problem is considered a little more deeply it will become evident that the values. H.T. and loudspeaker form a type of bridge circuil. This is. indeed. thie case. and provided each valve is taking equal current. as governed by the matching and biasing. the circuit will be in balance and no current will fow in the loudspeaker speech coil. This is the static D.C. condition.

However. when the valves are driven alternaiely by the signal. the condition of balance will be disturbed in sympathy with the signal and the difference current will flow through the speech coil. thus causing it to operate precisely the same as if it were connected to the secondary of a conventional output transformer. One of the most expensive and distortion-producing components has by this means been eliminated. but since a transformer is not available for matching purposes. the impedance of the loudspeaker requires to be greater than that of : loudspeaker used with a conventional output stage. in this connection. the speech ci, il impedance necessary is half the optimum load of a single valve.

The circuit in Fig. 4 illustrates how the tapped H.T. supply can be avoided. Since the signal "sees" a low impedance over the whole of the H.T. pattery. it matters little whether comection is-made at a centre-tap or at one end. The out-of-balance D.C. is blocked by capacitor C.
-The circuit at Fig. 2 show's clearly how the idea is extended to transistor circuits. Here a tapped bias source is adopted for $V 3$ and $\mathrm{V}+$. This is much a matter of convenience, though it does ease the balancing problem somewhat. since the tap is taken from the junction of wo seriesconnected 4.5 volt batteries (Ever Ready AB28s
(Contimed on pag? 3/3)



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or equivalents). which also serve to provide the required power for operating the Garrard rype BAI record player by was of the speed control potentiometer R14.

As already mentioned. the tho output transistors operate in class ${ }^{*} \mathrm{~B}$, ." or nearly so. In practice thes pass a small quiescent current so as to reduce distortion as the result of transistor non-linearity. Potentiometer networks comprising R9. R11 and R10. RI2 serve to bias the transistors in this way and also promote a small degree of negative feedhack and damping of resonances lormed by the transformer and transistor input capacitances. From our previous discussion. it will be clear that V3 and V4 are operated in the grounded-emitter mode.

Capacitor $\mathrm{C}_{7}$ smooths the effects caused by alternate "suitching" of the ourput transistors. while resistor R13 and capacior (4 form an overall negative feedback loop, giving some 10 dF of feedback at $400 \mathrm{c} / \mathrm{s}$. and provide a ift in bass response due 10 the frequency selective nature of the circuit. Capacitor (o is included to give enhanced stability al the higher frequencies.

Decoupling of the bias to $V / 1$ and $\backslash 2$ is secured by R7 and C3. as in ordinary valve circuits. The operating conditions of VI are slabilised by R3 and R 15 in conjunction with the emilter resistor R4, while the operating conditions of $V 2$ are similarly stabilised by R6. R. 5 and $R X$.

## Servicing Notes

As the operating conditions would be severely disturbed by shunting any netuorh resistor by a resistor of relatively low value and thus possibly result in permanent damage to a ransistor. voltage checks within the circuil must be made with a voltmeter of. at leas. 1000 ohms-per-volt sensitivity. Similarly. it is nol desirable to mahe or break connection within the circuil with the amplifier switched on. For evample, if a current check is required. the amplifier nust first be switched off. the circuit broken and milliammeter

inserted. The amplifier can then be switched on and the measurement made. The same applies. of course, on re-connecting the circuit.

The total " no-signal " current of the amplifier should range between 3.8 mA and 4.2 mA depending on the condition of the batterics. This is made up by current in the potentiometers. 0.4 mA of collector current in VI, 2 mA of collector current in V2 (taken between point A on L1 and the negative line) and $150 \mu \mathrm{~A}$ to $400 \mu \mathrm{~A}$ of collector current in V3 and V4. The out-olbalance loudspeaker current should not exceed $50 \mu \mathrm{~A}$. Taken on a 20,000 ohms-per-volt instrument. V'l collector should register negative 8.25


Fig. 4. --The H.T. tup can be aroided by the use of a coupling capacitor, $C$.
volts. V1 emitter negative 2 volts and $V 2$ cmitter negative 1.25 volts.

Transistors are very heat-conscious, and if overheated when operating are liable to destroy themselves in a very short time. They should thus be kept clear of soldering irons and other heat-producing devices. Soldering in and out of circuit should never be a prolonged process. A miniature low-power soldering iron is desirable. and even then a heat "sink" should be produced hy a pair of long-nose pliers held tight on the transistor connecting lead, between the point of soldering and the transistor itself. Moreover. the transistor leads should not be hent close to the seal.
Inadvertently reversing the supply polarity will almost certainly destroy a transistor. and this should also be borne in mind when making in sifu resistance checks with an ohmmeter. The negative terminal of most multi-range meters is usually in connection with the internal battery positive lead when used as ohmmeters-this is a point worth remembering.

When replacing components on the printed circuit, the faulty part should be snipped out so that sufficient connecting wire is available for the replacement part. The connecting leads of the new part should be tinned before attempting connection to the wire ends remaining above the board. In this way the heat period of soldering can be kept to a minimum.

reach 1958, the age of Kon and Eth. the culmination of all eartily passion and desire. the ne phus ultra of trembling trepidity and the incarnation of inconsequential insipidity-desire under the aspidistra-whose saga of frustration and impotence pursues us weekly.

I have. of course. referred to those cerebral offsprings of Frank Muir and Denis Norden. radioed by Dick Bentley and Jimmy Edwards and so lugubriously impersonated by Wallas Eaton and June Whitfield.
Whither wendest them: How will they perish? Only Messrs. Muir and Norden know this. So far as we are concerned. they follow the pattern set by famous predecessors such as Dan and Doris Archer or Dr. and Mrs. Dale-also Peter Pan: they never grow up. So far as we can see into the future, they can only terminate their existence by consummating their love which would. of course. transmogrify them completely. But Ron and Eth married-there would be crowds besieging Broadcasting House and the G.P.O. commandering fleets of auxiliary vans to carry the protest mail there: it would be unthinkable and unpardonable.
In the meantime may Eth's romantic illusions remain unshattered and Rons frightening timidity be continued for some time to come.

## Talking About Music

Antony Hopkins has made a name for himself as a conductor and broadcast critic which is wholly justified. His.. Sunday evening series. "Talking About Music." is quite delightful, very informative and equally refreshing to the novice or the devotee. Mr. Hopkins always has a happy simile ready and can illustrate passages on the piano with no mean dexterity. His programmes are most enjoyable.
Lady Godiva. ever since she rode the streets of Coventry in a costume that wouldn't pay a dividend for any silk or nylon merchant these days. has been a godsend to lyric writers, musichall jesters and dealers in ribaldry generally. A few years ago a famous judge had cause to remark that when he was a boy it took a sheep to dress a woman but. when he was speaking. it only required a couple of silkworms. The learned judge lived four or five hundrel years too late.
In "Scandal at Coventry"-specially written for radio-Clemence Dane treated the subject seriously and quite beautifully. The "ride," to the accompaniment of the slow clop/clop of hooves. sounded most effective and suggestive.

It provoked a thought: on ielerision the scene would vilher have been producd in an inhibited or meaningless manner or completely exorcised. but in "blind" radio it remained undistutbed. I'p radio!

## Famous Trials

The Thompson-Bywaters murder trial was one of the most lamous and exciting of all time So it was natural that it would find a place in the "Famous Trials" series: No. 4. in fact. It made both a dramatic plas and an historical docitmentary. Written by F. Temnyson-Jesse, commented on by Lord Birkett and produced by Nesta Pain. ail the poignancy and futility of the wretched allair were riade erystal clear. as was the savage brutality of sentencing Edith Thompson. Bary Wimbush played her with tragic simplicity and heartrending realism.

March 21st. 1918. That date means litte ar nothing to anyonc younger than the present writer; not so much. perhaps. as Waterloo or Agincourt. But to those of his own generation it ushered in a period during which neither Lloyd George nor Haig themselves knew whether the first world war was not. after all. going to be lost. The programme. "The March Retreat" (the title seems something like a masterpiece of understatement), visilly recalled those pulsating days. I hope it is followed up by one which 1 style in advancic "The Turn of the Tide and Victory." For it was less than eight months later that that frieluful war was all over and "in the bag."
As there was nothing new to say about it. nothing new was sought. But. with the Fifth Army Commatader hinself taking part. plus several important participants. it was a stern reminder. well presented. of events which will for ever remain in the memories of those even remotely commected will it.

## The ${ }^{\text {Lyons }}$

"Life With the tyons" proceeds pleasantly if not always with perfect tranquility! They make a delightful family---they are of course. very old friends mon-and it is always a pleasure to pop ill on them. The seripi maintains a reasonably high areage with very few wholly dull sessions. tike "The Archers." they will never grew up; if they did, that would be that?

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The Editor does not necessarily agree with opinions expressed by his correspondents

## Improving the "Soundmaster"

CIR,-The following suggestions will be found to improve the performance of the Soundmaster tape-recorder considerably:

1. Increase the value of the screen resistors of V1 and V2 to 330 h . The 33 k tesistors specified and supplied in the kit are incorrect and ruin the performance of the instrument. (The circuit diagram is correct.)
2. To prevent flutter and wow. especially with long-play tapes. the 500 ohm resistor in series with the take-off motor should be increased to at least 2k. (10w.). The tape will not spill and the resultant " slack" flutter is completely damped out by the pressure pads.
3. The pinch-wheelcam spring may be shortened by about $\frac{3}{8} \mathrm{in}$. to increase the pressure if there is any sign of slipping. though this should not be necessary.
4. The pressure pads should be applied as lightly as is consistent with good high-frequency response--J. H. Whiteley (GBAES) (Hıll).

## A Condenser Condition Tester

SIR.-Mr. J. Brown is apparently unaware of the excellent protection afforded by a metal rectifier shunted across a meter so as to give it a logarithmic scale. which is ideal for measuring the forming and leakage current of capacitors. The arrangement is shown in the accompanying circuit diagram. At very low voltages a me'al rectifier has considerable resistance and makes little effect on the lower readings of the meter. but as the voltage is increased the resistance of the rectifier decreases and so shunts a greater proportion of the current away from the meter. Two half-wave metal rectifiers are used connected as shown in order to give protection in both directions. and the resistance in series with the meter is adjusted to get fult-scale deflection at the current required.

A 1 mA meter shunted in this way can have a half-scale deflection of about -2 mA while reading 25 mA full scale, and for this a Westinghouse Type 2/6A rectifier. with its outside ends connected together. and a variable resistance of 2.500 ohms is suitable. Or the rectifier may be made up from an old copper-oxide one, using 6 discs for each section. no cooling fins being needed. A 5 mA meter can be used instead of the 1 mA . but the scale will be less open. The meter will need re-calibrating or a graph made,

and great accuracy cannot be cxpected since the resistance of the rectifier varies somewhat with temperature. but it is quite good enough for testing.

Using a single resistor (R13, Fig. I. April issue) to limit the current on all ranges is very unsatisfactory because if it is of sufficient value for the higher ranges it will grossly reduce the current at lower voltages. with misleading results. It is therefore best 10 put a limiting resistor in each range, except the lowest ones where there is sufficient resistance in circuit already to limit it to 25 mA .

II is interesting to note that. using the circuit shown, a 5 mA meter can be made to read 1 amp. and yet have a half-scale deffection of under 15 mA . For this a Westinghouse L.T.-7.A rectifier and resistance of about 130 ohms is required.-R. V. Ciorode (Tolland Bay).

## Intermittent Heater Circuit

SIR.-Occasionally an annoying fault occurs in A.C./D.C. sets which can prove difficult to trace. It is an intermittent heater. when the set has warmed up which results in all the valves having no heater current. The offending valve often will not show up when tested with no anode current taken. as it may not rise to the same temperature.

By a sketch. it can be seen that if a heater becomes o/c in a series heater chain. then the full mains volts are developed across the fault!

Safegnarding a me'ter. See Mr. R. V. Goode ${ }^{\circ}$ lener on the loft.

heater because with no heater current flowing there can be no voltage drop across the other resistances.

If a voltmeter is placed across a good value. then when the break oceurs the heater volts will fall to zero.

Therefore to determine which is the fault valve a roltage chech is made on each heater.
when one rises as the fault occurs this is the valic concerned.--Mons I.f (Islington).

## Using Old Eliminators

(IR. - In reply to H . Young (Reading). he may W be interested to know that I have been using an old hattery climinator as a very useful condenser leakage tester for many years. All that is needed is a neon lamp and holder and a bracket to fix it on top of the eliminator: then fix two flexible leads in the H.T. positive and negative sockets with the neon in series with one of them. and terminated in spring clips.

In use 1 find that condensers down to $.005 \mu \mathrm{~F}$ give a useful indication on first switching on. leaky condensers are immediatoly revealed by a steady glow or continuous Hashing. High value resistors. which may give no indication on an ohmmeter may also be readily tested for continuity.

I hope the above hint mas be useful to Mr. Young and other readers. -H. E. Howard (Bournemouth).

## Tape Recorder Improvement

GIR. Almost all tape recorders hate at least two heads. record-play. and erase.
The record-play head is in use all the time. but the erase head stands idle when playing back. or recording additions to a previous recording. This is obviously a waste of perfectly good head, and the following notes show how it mas be used as a monitoring head and for giving echo effects. The illustration shows the sysiem used. The signal from the low-impedance erase

## Aligning Alba Models 3211, $6221,6231,6241$ and 6251

THE following are the trimming details for the alignment of the above reccisers which were dealt with in our April isste.

Retune the generator to 154 Kc , s and the receiver to 1,450 metres and adjusi $L .15$ and $L X$ for maximum output. Alter tuning to $250 \mathrm{Kc} / \mathrm{s}$ (1.200 metres) and adjun (Xo and CX3 for maximum output.

Finally, the S.W. band: Tune the receiver to 50 metres and the generator $106 \mathrm{Me} s$ and adjust 1.13 and L6 for maximum ouput. Retune to 17 metres ( $17.65 \mathrm{Mc} / \mathrm{s}$ ) and adjust ( $X+$ and CXI for maximum output. For best overall results the operations should be repeated for optimum tracking and sensitivits.

## F.M. Alignment

Switch the recciver to F.M. and tune to the low-frequency end of Band II. Adjust the R.F. and oscillator trimmers $C 5$ and ( 4 to the centre of their range. Connect a valve voltmeter between the junction of C7 and C 8 on the F.M. tumer and adjust clo lor minimum reading. ohserving only the dip in meter indication which occurs when C 10 is nearest its minimum value.

Conneet a microammeter and series 200 K resistor berneen pin 2 of V4 and chassis, with the positive
head is fed to a high gain transistor amplitier. The circuit shown gives a gain of about 300 and takes 0.8 mA from the 250 s . supply. The output is suitable for feeding to a valve amplifier or to high-impedance headphones. Other types wl transistor may be used if care is taken not to exceed their permitted voltage and curremt ratings. The spare contacts on must push-button tape decks cat be arranged to give the suitching arrangements $S$.


The tape recomder swithing arrangement described ly Mr. Dobron.

The quality of the output will depend on the erase head gap and the tape speed. Some top cut is inevitable with wider gaps. The device is most suitable for monitoring. when adding to recordings already on the tape but interesting echo effects nus also be obtained by feeding the monitoring signal from the erase head amplifier back to the record head.--J. Dobson (Sheffield).
neter terminal to chassis, apply a 10.7 Me;s unmodulated signal across 1.5 and adjust 1.18 . 1.12 and LII for maximum indication on microammeter. Make a note of this reading. Transfer negative meter connection "to point. "T" on circuit diagran and carefully adjust the core in 1.19 for a meter reading exactly half of that noted above. Repeal all the operations given in this paragraph.

Reconnect the negative meter lead to pin 2 of V4. inject a $10.7 \mathrm{Me} / \mathrm{s}$ unmodulated signal to the tuner by way of a tight loop of wire around the ECCX5 and adjust 1.4 and L. 5 for maximum indication on the meter. It is recommended that a chech of the F.M. I.F. tesponse is made at this point by detuning the generator $100 \mathrm{kc} / \mathrm{s}$ either side of the $10.7 \mathrm{Mc} / \mathrm{s}$ (plus and minus $100 \mathrm{Kc} / \mathrm{s}$ ) and checking that the reduction in meter indication is no greater than 40 per cent. (.3 db).

Set the receiver tuning to the calibration marh at the low frequency of Band 11 . injeet into the Band 11 acrial terminals a signal of 87.9 Mess. deviated at $25 \mathrm{kc} / \mathrm{s}$. and adjust ( 9 and (5 for maximum indication on an output meter connected to the evtension loudspeaker sockets. Tane the receiver to $95 \mathrm{Mc} / \mathrm{s}$ and retune the generator to coincide. Adjust the spacing between ( 7 and C8 for calibration at this frequency.

Disconnect meter and signal generator. conneet F.M. aerial and adjust (il for optimum sensitivit!. Seal all trimmers lightly with was.
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"SOUND 777 " PRICE INCREASE

OWING to the increase in the list price of the Collaro tape transcriptor trom $£ 22$ to $£ 25$. Tape Recorders Ltd., have found it necessary to amend the price of the "Sound 777 " tape recorder from 40 gns. to 44 gns . list. They have also been faced with the additional rising costs of both labour and material and the total increases have been taken into consideration in fixing this new price.

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## THE MOTEK K9 TAPE DECK

$W^{E}$ again show this deck which. unfortunately: was incorrectly described in the caption in our April issue as The New Motor Tape Deck. The manufacturers, who submit each deck to exhaustive


The new Motek tape deck.
lests before despatch guarantee them for a period of 12 months and we understand that production of the Motek K9 is now in full swing to meet the ever increasing trade demand.--Modern Techniques, Wedmore Street, London, N. 19.

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## CRYSTAL SETS

2／－each
1937 Crystal Receiver ．．．PW7I＊
The＂Junior＂Crystal ．．．PW94＊
$2 / 6$ each
Dual－Wave＂Crystal Diode＂

Crystal
PW95＊

## STRAIGHT SETS

## Battery Operated

One－valve：2／6 each
The＂Puramid＂One－ valver（HF Pen）
The Modern One－ valver
wo－valve ： $2 / 6$ each
The Signet Two（D \＆
LF)

3 ＇6 each
Modern Two－valver（two band receiver，
Three－valve ： $2 / 6$ each
Summit Three（HF，Pen． D，Pen）
The＂Rapide＂＂Straight 3 （ $\mathrm{D}, 2 \mathrm{LF}$（RC \＆ Trans）
F．I．Camm＂s＂Sprite＂ Three（IIF，Pen．D）． Te1）

PW93＊
PW96＊

PW76＊

PW98＊

PW37＊

PW82＊

3／6 each
The All－dry Three
．．．PWソ7＊
Four－valve ： 26 each
Fury Four Super（SG． SGi，D，Penl ．．．

PW34C＊

## Mains Operated

Two－valve： 2 ＇6 each
Selectone A．C．Radio－ gram Two（D，Pow）．．．
Three－valse：4／－each
A．C．Band－Pass 3
PWI9＊

Four－valve ： $2 / 6$ each
A．C．Fury Four（SG，SG，
D．Pen）
A．C．Hall－Mark（HF． Pen，I），Push Pall）．

## SUPERHETS

Battery Sets： $2 / 6$ each
F．J．Cammis 2－valve Superhet

PW52＊
Mains Operated：4／－each
＂Coronet＂A．C．4 ．．．PW 100＊
AC／DC＂Coronet＂Four PW 101＊＊

No．of Blueprint

## SHORT－WAVE SETS

| Battery Operated |  |
| :---: | :---: |
| One－valse ： $2 / 6$ each |  |
| Simple S．W．One－valver | PW88 |
| Two－valve： $2 / 6 \mathrm{each}$ |  |
| Midget Short－wave Two （D，Pen） | PW38A＊ |
| Three－valve ： $2 / 6$ each |  |
| ，Experimenter＇s Short－ Wave Three（ $\mathbf{S C}, \mathrm{D}$, Pow） |  |
| The Prefect 3 （D， 2 LF （RC and Transi） | ．PWg |
| The Band－spread S．W Three（hff，Pen，D （Pen），P＇cn） | PW68＊ |
| MISCEILIANEOU | S |
| 2＇6 each |  |
| S．W．Converter－Adapter |  |
| The P．W．3－speed Auto－ gram ．．．．．．（2 slects），8－＊ |  |
| The P．W．Monophonic <br> IFectronic Organ（2 sheets）， 8 － |  |
| TELEVISION |  |
| The＂Argus＂（6in．C．R．Tube），3＇＊ |  |
| The＂Simplex＂ |  |
| The P．I．Band III Converter | reter 16＊ |

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to utilise old components which then
man hate in fmelr spares bor．The
molority of the components for the：e
rectevers are no longer stacked bu
retallers．

AMATEUR WIRELESS AND WIRELESS MAGAZINE STRAIGHT SETS

Battery Operated
One－value：2／6
B．B．C．Special One－ valver $\quad$ ．．．$\ldots$ AW387＊

## Mains Operated

Two－valse： 26 each
Consoelectric Two（D， Pent．A．C．

## SPECIAL NOTE

THESE blucwrints are drown fu？ size．The Jisucs rontalining deseriptions of these sets are mow dut of print．but an asterisk demote that coustuructional details are abit： able．free with the biuepront．

The index hatiens which procegle the Blufprint Number inuleate ther pertot （c） 11 which the description ap，eear： Thus 1．W．refers to PRACTIC +L W＇IRFIESS．A．W．to Amateur Whetes． W．M．to Wirelese Mrgazine．

Send（preterably）a postal order to cover the cost of the Blueprins stamps uver Gid．unacceptable）to 1＇RACTICAL WIRELESS，BIいのかire Hept．Giforge Newnes．Latd．「 「uwter Houke．Southampton street，Strant， W．C． 2 ．

No．of
Bliteprine

## SHORT－WAVE SETS

## Battery Operated

One－valve ： $2 / 6$ each
S．W．Onc－vilver for American

AW429＊
Fwo－valve ：2／6each
Ultra－short laatery Two
（SGi，det Pen）．．．．．．
WM402
Four－valve： 36 each
1．W．Short Wave World－ heater（HF，Pen，D，RC， Trans）

AW4．36＊
Standard Four－valver
Short－watver（SG，D．
I－I－ 1 ）
WM383＊

## Mains Operated

Four－valve： $3 / 6$
Standard Four－valve A．C．
Short－waver（SC．1）．
R（．Irans）．．．．．．WM391＊

## MISCELLANEOUS

Fombustast＇s Power Am－ plitier（ 10 Watls）$(3 ; 6)$ WM387＊

Listener＇s 5－watt A．C．
Amplidier（3／6）．．．WM．392＊
De Luxe Concert A．C．
Elecirogram（2，6）．．．WM403＊

## 

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* Highly sensitive.
* Attractive lightweight contemporary case.

Car Radio Conversion Components. 8/- extra.

| We can supply all |
| :--- |
| these items including |
| Cabinet for fillio.0. |
| P.P. 2/6. All parts |
| sold separately. |
| Circuit diagram and |
| shopping list free. |



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