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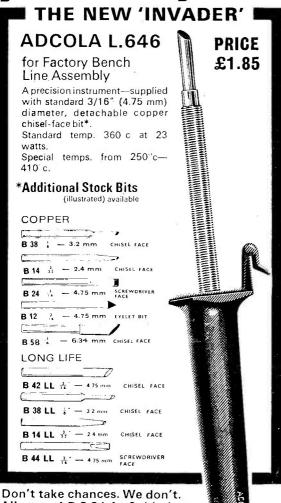
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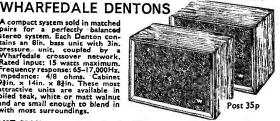
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BSR McDONALD MP60

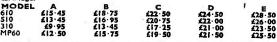
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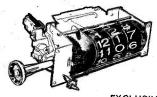
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THE RADIO **CONSTRUCTOR**

September Issue

Constructional Articles include

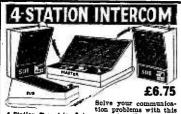
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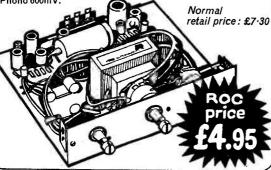
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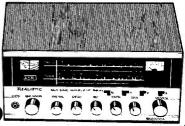
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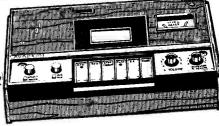
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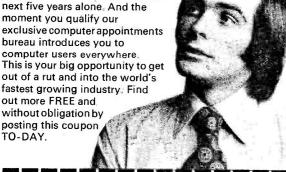
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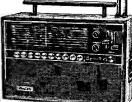
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SK1 SOLDERING KIT

In rigid plastic "tool box" containing Model CN - 15 watts - 240 volts miniature iron fitted $\frac{3}{16}$ " bit. Spare bits $\frac{5}{32}$ " and $\frac{3}{32}$ ". Reel of resin-cored solder, heat sink, cleaning pad, stand and booklet "How to Solder".



SK2 SOLDERING KIT

In polystyrene pack, containing 15 watt miniature soldering iron, 240 volts fitted with $\frac{3}{16}$ " bit, 2 spare bits $\frac{5}{32}$ " and $\frac{3}{32}$ ". Coil of resin-cored solder, heat sink, 1A fuse and booklet "How to Solder".

£2.40

£2.75



- 15 watts
- 240 voits

£1.70

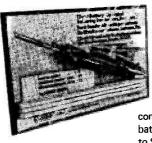
Fitted with nickel plated $\frac{3}{32}$ " bit and packed in handy transparent box.



ES240 D 25 watt soldering iron

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In transparent display pack, fitted with long life iron-coated bit $\frac{1}{8}$ " diam. Interchangeable spare bits $\frac{3}{32}$ ", $\frac{3}{16}$ ", $\frac{1}{4}$ " (extra) available. Improved design to ensure strong and reliable high speed iron. Heats up in 2 minutes.



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Battery-operated
12 volt
soldering iron

Complete with 15 ft (4-50m) lead, 2 heavy gauge clips for instant

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All cabinets are new and carefully designed acoustically with speakers mounted on \(\frac{1}{2}\)in. chipboard baffles. All speakers are ex-TV high quality with hi-flux magnets carefully matched and tested.

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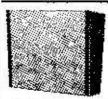


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Simple unit connects direct to amplifier and speakers to give attenuated headphone output, has 2 position switch to give headphones only—speakers only.
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II" \times 9½" \times 5" deep, tapering to 3½" deep covered in Black reseine and Mottled fawn Vynair £2·10 P & P 25p.

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Wedge shaped extension speaker 7½ x 6½ x 4in.

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Keyhole slot at back. Fitted with 3 ohm speaker unit. Only £1.25. P. & P. 36np each

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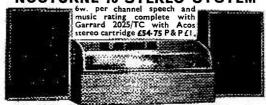


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1T4	-16	30C15	-63	DY802	-40	EM81	•41	PCL84	•85	UBC41	-52
384	-26	30C17	-80	EABC80	-82	EM84	-32	PCL85	•42	UBF80	-84
3V4	•37	30C18	-68	EAF42	-50	EM87		PCL86		UBF89	•88
5U4G	-26	30F5	•70	EB41	-40	EY51	-38	PCL88		UCC84	-88
5V4G	-85	30FL1	-68	EB91	-11	EY86	-81	PCL80		UCC85	-86
5Y3GT	-28	30FL12	•70	EBC33	-40	EZ40	-43	PENA	4 .77	UCF80	-84
5Z4G	-85	30FL14	-70	EBC41	-54	EZ41	-48	PEN36	C -70	UCH42	-62
6/30L2	-58	30L1	-80	EBC90	-22	EZ80		PFL20	0 .58	UCH81	-82
6AL5	-11	30L15	•59	EBF80	-82	EZ81	-23	PL36	-49	UCL82	-84
6AM6	-18	30L17	•71	EBF89	-81	GZ30	-35	PL81	-46	UCL83	-55
6AQ5	-26	30P4	-65	ECC81	-17	GZ32	-42	PL81A		UF41	-56
6AT6	.22	30P12	•77	ECC82	-20	GZ34	-50	PL82	-81	UF89	-81
6AU6	.21	30P19	-65	ECC83	-35	KT41	•77	PL83	-83	UL41	-60
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6BW7	.55	30PL15	-90	ECF82	-26	LN329	.72	PM84		UY41	-40
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6F14	-42	35W4	-25	ECH42	-61	N78	-87	P¥32	- 55	VP4B	.77
6F23	.70	35Z4GT	-25	ECH81	.29	P61	-45	PY33	-55	Z77	-22
6F25	-60	807	-45	ECH83	-40	PABC80	-34	PY81	-27	Transit	
6K7G	-12	6063	-62	ECH84	-86	PC86		PY82	-25	AC107	.17
6K8G	.17	AC/VP2	•77	ECL80	-82	PC88		PY83	-28	AC127	·18
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6SN7GT	-80	B729	-62	ECL86	-36	PC97	·40	PY800		AF115	-20
6V6G	-23	CCH35	-67	EF39	-88	PC900	-33	PY801	-84	AF116	-20
6V6GT	-31	CY31	-82	EF41	-60	PCC84	-80	R19	-80	AF117	-20
6X4	-28	DAF91	.22	EF80	-23	PCC85	-27	R20		AF118	-48
6X5GT	.28	DAF96	-36	EF85	-28	PCC88	-42	U25	-65	AF125	•17
7B7	-88	DF33	-88	EF86	-81	PCC89	•46	U26		AF127	.17
10P13	-58	DF91	-16	EF89	-26	PCC189	-48	U47	-65	OC26	-25
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12AU6	-22	DK32	.87	EF184	-31	PCF82	.81	U52	-31	OC71	-12
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12AX7	-23	DK92	.42	EL33	-55	PCF800	-67	U191	-60	OC75	.12
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20P3	-80	DL92	-26	EL84	-23	PCF805	-63	U301	-45	OC82	.19
20P4	.92	DL94	.37	EL90	-26	PCF806	-58	U329	-66	OC82D	12 12
25L6GT											
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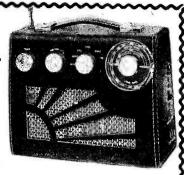
NEW! ROAMER 10 WITH VHF INCLUDING AIRCRAFT

10 TRANSISTORS. 9 TUNABLE WAVEBANDS, MW1, MW2, LW, SW1, SW2, SW3, TRAWLER, BAND, VHF AND AIRCRAFT BAND

Built in Ferrite Rod Aerial for MW/LW. Retractable, chrome plated 7 section Telescopic Aerial, can be augled and rotated for peak short wave and VHF listening. Push Pull output using 600mw Transistors. Car Aerial and Tape Record Sockets. Switched Earplece Socket complete with Earplece. 10 Transistors plus 3 Diodes. 7in x 4 in Speaker. Air Spaced ganged Tuning Condenser with VHF section. Volume on/off, Wave Change and Tone Control. Attractive Case in black with silver blocking. Size 9" x 7" x 4". Easy to follow instructions and diagrams. Parts price list and easy build plans 30p (CFREE with parts). (FREE with parts).

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P. P. & Ins. 50p (Overseas P. & P. £1)



ROAMER FIGHT MkI NOW WITH

SWITCHED TONE CONTROL

7 Tunable Wavebands: MWI, MW2, LW, SWI, SW2, SW3 and Trawler Band. Built in Ferrite Rod Aerial for MW and LW. Retractable chrome plated Telescopic serial for Bndv Waves. Push pull output using 600mW transistors: Car aerial and Tape record sockets. Selectivity switch. Switched earpiece socket complete with earpiece. St transistors plus 3 diodes. 7in. × 4in. Speaker. Air spaced ganged tuning condenser. Volume/on/off, tuning, wave change and tone controls. Attractive case in rich chestant shade with gold blocking. Size 9 × 7 × 4in. approx. Easy to follow instructions and diagrams. Parts Price List and Easy Build Plans 25p (FREE with parts).

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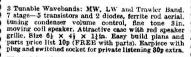
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Heavy construction. Latest high efficiency ceramic magnets. Treated Cone surround. "D" indicates Tweeter Cone providing extended frequency range up to 15,000 c.p.s. Impedance 3 or 8-15 ohms. Please state choice. Exceptional performance at low cost.

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with 10in. Hi-Fi 'speaker and Tweeter.

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Size 25 x 19 x 10in.

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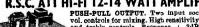
R.S.C. TA6 6 Watt HI-FI SOLID STATE AMPLIFIER

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200-250v. AC mains operated. Frequency Response 30-20,000 c.p.s. —2dB. Harmonic Distortion 0.3% at 1,000 c.p.s. Separate Bass and Trabe. Input selector switch. Output rating I.H.P.M. Fully enclosed enamelled case, 9\frac{9}{4} \times 2\frac{4}{4} \times \frac{1}{4} \times \

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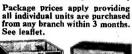
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SPECIFICATION R100/101

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U24 20 Germanium 1-Amp rectifiers GJM up to 300 PIV 0-50 U25 25 300Mc/s NPN silicon transistors 2N708, B8Y27 0-50 U26 30 Fast switching silicon diodes like IN914 micro-min 0-50 U28 Experimenters' assortment of integrated circuits, untested.	U21	30 A.F. germanium alloy transistors 2G300 series & OC71	0.50
U25 25 300Mc/s NPN silicon transistors 2N708, B8Y27	U23	30 Madt's like MAT series PNP transistors	0-50
U26 30 Fast switching silicon diodes like IN914 micro-min 0.50 U28 Experimenters' assortment of integrated circuits, untested.	U24	20 Germanium 1-Amp rectifiers GJM up to 300 PIV	0-50
U28 Experimenters' assortment of integrated circuits, untested.	U25	25 300Mc/s NPN silicon transistors 2N708, BSY27	0.50
	U26	30 Fast switching silicon diodes like IN914 micro-min	0.50
	U28	Experimenters' assortment of integrated circuits, untested. Gates, flip-flops, registers, etc., 8 assorted pieces	1.00

10 1-Amp SCR's TO-5 can up to 600 PIV CR81/25-600 1-00

20 Sil. Planar NPN trans, low noise amp 2N3707 0-50 25 Zener dlodes 400mW D07 case mixed volts, 3-18 0.50

20 Fast switching all. trans. NPN, 400Mc/s 2N3011 0-50

30 RF germ. PNP trans. 2N1303/5 TO-5 0-50

10 Dual trans. 6 lead TO-5 2N2060 0-50

25 RF germ. trans. TO-1 OC45 NKT72 0-50

10 VHF germ. PNP trans. TO-1 NKT667 AF117..... 0-50 Code Nos. mentioned above are given as a guide to the type of device in the Pak. The devices themselves are normally unmarked.

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1129

U31

U32

BT27

U38

TT39

1741

U42

BRAND	NEW	TEXAS
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Coded an	d Guar	anteed
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T1	8 2G371.	A OC71
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Г3	8 2G3744	
Г4	8 2G381.	A OC81
T5	8 2G382	
T 6	8 2G344.	A OC44
T7	8 2G345	
T8	8 2G378	OC78
T 9	8 2G399.	A 2N130
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Q35	3 Sil. PNP TO-5 2 × 2N2904 & 1 × 2905	
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Q38	7 NRV 4-2 4 2000004 2 2000000	0.50
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BP70s = 7400 BP10s = 7401 BP10s = 7410 BP10s = 7440 BP10s	1979.00 7409	(with open collector output)			
BP70s = 7400 BP10s = 7401 BP10s = 7410 BP10s = 7440 BP10s	BP02 = 7402 $BP03 = 7403$	Quad 2-input pos. NAND gates	0-15	0.14	0.12
BP70s = 7400 BP10s = 7401 BP10s = 7410 BP10s = 7440 BP10s	****	(with open collector output)			
BP10 = 7410	BP04 = 7404 BP05 = 7405	Hex inverters	0.15	0.14	0.12
		output)	0.15	0.14	0.12
BP46 = 7446	BP10 = 7410	Triple 3-input pos. NAND gates	0.15		0.12
BP46 = 7446	RP20 = 7420	Dual 4-input scamitt trigger	0.29	0.26	0.24
BP46 = 7446	BP30 = 7430	8-input pos. NAND gates	0.15	0.14	0.12
BP46 = 7446	BP40 = 7440	Dual 4-input pos. NAND buffers	0.15		
BP46 = 7446 BCD-to-seven-segment decoder/driver BCD-seven-segment decoder/driver BCD-seven-segment decoder/driver BCD-seven-segment decoder/driver BCD-seven-segment decoder/driver BCD-seven-segment decoder/driver BCD-to-seven-segment decoder/driver BCD-to-seven-segment decoder/driver Decoder-segment decoder-driver Decoder-driver Decoder-segment decoder-driver	BP41 = 7441	BCD to decimal nixie driver	0.87		
BP46 = 7446 BCD-to-seven-segment decoder/driver BP47 = 7447 BCD-seven-segment decoder/drivers (15% outputs) BP48 = 7448 BCD-to-seven-segment decoder/drivers (15% outputs) BP50 = 7450 BCD-to-seven-segment decoder/driver BP50 = 7450 BCD-to-seven-segment decoder/driver BP51 = 7451 BCD-to-decimal decoder/driver BP51 = 7451 BCD-to-decimal decoder/driver BP51 = 7451 BCD-to-decimal decoder/driver BP52 = 7452 BCD-to-decimal decoder/driver BP53 = 7452 BCD-to-decimal decoder/driver BP54 = 7454 BCD-to-decimal decoder/driver BP54 = 7455 BCD-to-decimal decoder/driver BP54 = 7456 BCD-to-decimal decoder/driver BP54 BCD-to-decimal decoder/driver BCD-to-decimal decod	BP42 = 7442	BCD to decimal decoder (4-10 lines,	0.00	0.04	
BP48 = 7448 BCD-seven-segment decoder/drivers 0-97 0-94 0-88 BP50 = 7450 Expandable dual 2-input and-or-invert 0-15 0-14 0-12 0-15 0		BCD-to-seven-segment decoder/driver			
BP48 = 7448 BCD-to-seven-segment decoder/driver 0.97 0.94 0.88 0.97 0.97 0.98 0.97 0.98 0.97 0.98 0.98 0.97 0.98 0.98 0.97 0.98	BP47 = 7447	BCD-seven-segment decoder/drivers			
BP50 = 7450	BP48 - 7448	BCD-to-seven-segment decoder-deliver			
December 2-15		Expandable dual 2-input and-or-	0.87	0.94	0.88
## BP55 = 7453 Gates College C		invert	0.15	0.14	0.12
BP55 = 7453	BP51 = 7451		0.15	0.14	0.10
BP54 = 7454	BP53 = 7453	Quad 2-input expandable and-or-			
BP66 = 7460 Dual 4-input expander 0.15 0.14 0.12 BP70 = 7470 Single-phase J. K filp-flop 0.29 0.28 0.24 BP72 = 7472 Master slave J. K filp-flop 0.29 0.28 0.24 BP73 = 7473 Dual Master slave J. K filp-flop 0.37 0.35 0.32 BP74 = 7475 Qual Master slave J. K filp-flop 0.37 0.35 0.32 BP76 = 7475 Qual latch 0.47 0.45 0.42 BP76 = 7476 Gatef full adders 0.47 0.45 0.42 BP80 = 7480 Gatef full adders 0.47 0.45 0.42 BP81 = 7481 16-bit read/write memory 0.97 0.94 0.88 BP81 = 7482 2-bit binary full adders 0.97 0.94 0.88 BP83 = 7482 Quad full adder 0.97 0.94 0.88 BP83 = 7482 Quad full adder 0.97 0.94 0.88 BP90 = 7400 8-bit shift register 0.97 0.94 0.58 BP91 = 7401 8-bit shift register 0.97 0.94 0.78 BP93 = 7493 4-bit binary full shift register 0.97 0.94 0.78 BP94 = 7494 4-bit binary full shift register 0.97 0.46 0.58 BP95 = 7496 Constant of the product	BP54 - 7454	invert			
BP70 = 7470 BP72 = 7472 BNster slave J-K filp-flop	BP60 = 7460	Dual 4-input expander	0.15	0-14	0-12
Depth	BP70 = 7470	Single-phase J-K flip-flop	0.29	0-26	0.24
Depth	BP72 = 7472	Master slave J-K flip-flop	0.29		0.24
Depth		Dual Master slave J-K flip-flop		0.35	
BP76 = 7476 BP80 = 7480 Gated full adders		Out D type mp-nop			
Section Sect	BP76 = 7476	Dual J-K with pre-set and clear			
Section Sect	BP80 = 7480	Gated full adders	0.67		
Section Sect	BP81 = 7481	16-bit read/write memory	0.97	0-94	0-88
Section Sect	BP83 = 7483	Quad full adders	0.97		
Divine Opt welve counters 0.67 0.64 0.58	BP86 = 7486	Quad 2-input exclusive NOR sates	0.32		
Divine Opt welve counters 0.67 0.64 0.58		BCD decade counter	0.67	0.64	
BP100	BP91 = 7491	8-bit shift registers		0.84	0.78
BP100	BP92 = 7492 BP93 = 7493	Jivine-by-twelve counters			
BP100	BP94 = 7494	Dual entry 4-bit shift register	0-77		
BP100	BP95 = 7495	4-bit up-down shift register	0.77		
BP100 = 74100 BP104 = 74100 BP105 = 74100 BP106 = 74105 BP106 = 74105 BP106 = 74105 BP107 = 74107 BP108 = 74107 BP108 = 74107 BP109 = 74107 BP110 = 74110 BP1110 = 7	BP96 ≈ 7496	o-oit paranei in paranei out shift-	0.77	0.74	0.00
BP104 = 74104 Single J-K filp-flop equivalent 9000 Series Single J-K filp-flop equivalent 9001 Series Single J-K filp flop equivalent 9001 Series	BP100 = 74100	8-bit bistable latches			
BP105 = 74105 Single J-K flip flop equivalent 9001	BP104 = 74104	Single J-K flip-flop equivalent 9000			
BP107 = 74107 Dual Master slave flip-flops 0.97 0.94 0.88 BP110 = 74110 Dual Master slave flip-flops 0.40 0.38 0.36 BP110 = 74110 Dual Master slave flip-flops 0.55 0.52 0.50 BP111 = 74111 Dual data lock-out flip-flop 1.25 1.15 1.00 BP118 = 74118 Hex set-reset latches 1.00 0.95 0.90 BP119 = 7410 Hex set-reset latches 2.4-pin 1.35 1.25 1.10 BP121 = 74121 Monostable multivibrators 0.67 0.64 0.58 BP141 = 74143 BCD-to-decimal decoder/driver 0.67 0.64 0.58 BP145 = 74145 BCD-to-decimal decoder/driver 0.67 0.40 1.30 BP150 = 7450 6.50 data selector 1.80 1.70 1.60 BP151 = 74151 S-bit data selectors (with strobe) 1.00 0.95 0.90 BP163 = 74453 Dual 4-line-to-l-line data 1.20 1.10 0.95	BP105 = 74105		0.87	9.94	0-88
BP141 = 74141 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74150 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74150 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP151 = 74151 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP151 = 74151 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 1-40	DD707 - 24107	series			
BP141 = 74141 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74150 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74150 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP151 = 74151 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP151 = 74151 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 1-40	BP110 = 74107	Gates master-slave flip-flops		0.38	
BP141 = 74141 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74151 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74150 16-bit data selector 1-80 1-70 1-60 BP151 = 74151 B-bit data selector 1-80 1-70 0-95 0-90 0-90 BP153 = 74153 Dual 4-line-to-l-line data 1-20 1-10 0-95 0-90 1-70	BP111 = 74111	Dual data lock-out flip-flop			
BP141 = 74141 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74151 BCD-to-decimal decoder/driver O/C 1-50 1-40 1-30 BP150 = 74150 16-bit data selector 1-80 1-70 1-60 BP151 = 74151 B-bit data selector 1-80 1-70 0-95 0-90 0-90 BP153 = 74153 Dual 4-line-to-l-line data 1-20 1-10 0-95 0-90 1-70	BP118 = 74118	Hex set-reset latches	1.00	0.95	0.90
BP140 = (4145) BUD-to-decemial decoder/driver O/C 1.50 1.40 1.30 BP150 = (4140 16-bit data selector 1.80 1.70 1.60 BP151 = 74151 S-bit data selectors (with strobe) 1.00 0.95 0.90 8P183 = (41453 Dual 4-line-to-l-line data 1.20 1.10 0.95	BP119 = 74119 BP191 = 74191	Monostable multivibrators	1-35	1.25	1.10
BP140 = (4145) BUD-to-decemial decoder/driver O/C 1.50 1.40 1.30 BP150 = (4140 16-bit data selector 1.80 1.70 1.60 BP151 = 74151 S-bit data selectors (with strobe) 1.00 0.95 0.90 8P183 = (41453 Dual 4-line-to-l-line data 1.20 1.10 0.95	BP141 = 74141	BCD-to-decimal decoder/driver	0.67		
BP151 = 74151 8-bit data selector	BP145 = 74145	BCD-to-decimal decoder/driver O/C	1.50	1.40	1.30
Date 1 110 089	BP150 = 74150 BP151 = 74151	0 1 14 1-41-4 1 111 1 1 1	1.80		1.60
Date 1 110 089		Dual 4-line-to-1-line data	1-00		
BP155 = 74155 Dual 2- to 4-line decoder 0.C 1.40 1.80 1.20 BP156 = 74156 Dual 2- to 4-line decoder 0.C 1.40 1.80 1.20 BP160 = 74150 Sync. decade counter 1.80 1.70 1.60 BP161 = 74151 Sync. 4-bit binary counter 1.80 1.70 1.60 BP190 = 74190 Sync. up-down BCD counter 3.50 2.00 2.00	BP154 = 74154	The state of the charge of the state of the	1.80		
Br130 = 74166 Dual 2- to 4-line decoder O/C . 1.40 1.80 1.20 BP160 = 74160 Spic. decade counter . 1.80 1.70 1.60 BP161 = 74161 Spic. 4-bit binary counter . 1.80 1.70 1.60 BP190 = 74190 Spic. up-down RCD counter . 3.50 3.50 3.60	BP155 = 74155	Dual 2- to 4-line decoder	1.40	1.30	
BP161 = 74161 Sync. 4-bit binary counter	BF100 = 74156 RP160 = 74160	Dual 2- to 4-line decoder O/C	1.40		
BP190 = 74190 Sync. up-down BCD counter 3.50 2.08 2.00	BP161 = 74161	Sync. 4-bit hinary counter		1.70	
	BP190 = 74190	Sync. up-down BCD counter	3.50	3-25	3.00
BP191 = 74191 Sync. binary up-down counter (single clock line)	BP191 = 74191	Sync. binary up-down counter (single			
BP192 = 74192 Sync. Hy-down decade counter 2.70 3.95 3.00	BP192 = 74192	Sync. Hu-down decade counter			
BP193 = 74193 Sync. bluary up-down counter (tow		Sync. bluary up-down counter (tow			
clock lines)	RP106 - 7410c	Clock lines)	2-10	1.95	
BP197 = 74197 Pre-setable 50MHz binary counter 1.80 1.70 1.80	BP197 = 74197	Pre-setable 50MHz binary counter	1.80	1.70	
	BP198 = 74198	8-bit parallel L-R shift register	5.50		
BP198 = 74198 8-bit parallel L-R shift register 5-50 5-00 4-00 BP199 = 74199 8-bit parallel access shift register 5-50 5-00 4-00	Br199 = 74199	8-bit parallel access shift register		5.00	4.00

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$UIC03 = 12 \times 7403N$		12 × 7460N 50p	$UIC86 = 5 \times 7486N 50p$
$UIC04 = 12 \times 7404N$		8 × 7470N 50p	$UIC90 = 5 \times 7490N500$
$UIC05 = 12 \times 7495N$		8 × 7472N 50p	$UIC92 = 5 \times 7492N 50p$
$OIC10 = 12 \times 7410N$		8 x 7473N 50p	$UIC93 = 5 \times 7493N 50p$
$UIC20 = 12 \times 7420N$		8 × 7474N 50p	$UIC94 = 5 \times 7494N 50p$
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BP933	Dual 4-input expander				13p	12p	īî
BP935	Expandable Hex Inverter				13p	120	îî
B P936	Hex Inverter				13p	12p	11
BP944	Dual 4-input NAND expans		huffer w	thou	Lup	LLy	11
	pull-up				13p	12p	11
BP945	Master-slave JK or RS						
BP946	Quad, 2-Input NAND.				25p	24p	22
BP948		114	963		12p	11p	10
	Master-slave JK or RS	E > 0	"atte		25p	24p	22
BP951	Monostable				65p	60p	55
BP962	Triple 3-input NAND				12p	11p	10
3 P 9093	Dual Master-slave JK with	separ	ate clock	k	40p	38p	35
3P9094	Dual Master-slave JK with	separ	ate clock	τ	40p	38p	35
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PRACTICAL WIRELESS

VOL 47 NO 6

Issue 776

OCTOBER 1971

Who makes what?

WELL, of course, that's easy! When you buy a radio set, hi-fi amplifier or piece of electrical equipment, the trade mark tells you that. Or if the mark is an unfamiliar one you can find the piece moulded on the cabinet or printed on the carton which says 'Made in Outer Mongolia' or whatever.

People are interested in where products are made for a variety of reasons. The possible standard of workmanship, value-for-money, reputation, reliability, the position of after-sales service, availability of spare parts—or even personal political motives—are all valid reasons why a purchaser would want to know who makes what, and where.

From November 30th, however, the Merchandise Marks Act comes to an end and from that date it will no longer be obligatory for imported goods to bear a 'country-of-origin' mark. The casual customer will therefore not be sure if he is buying a radio set made in Birmingham or Baluchistan, or an amplifier which is a Chinese-copy of an American design made in Hong Kong or Japan.

Of course, there will still be trade marks which proclaim that the goods are true-blue British or come from well-respected foreign companies. Sadly, though, things are not always what they seem for many tape recorders, radio and TV receivers, bearing the names of old-established British manufacturers have for some time been made overseas and imported for the home market.

From November 30th, protection of the consumer has been thown at the retailer who, when asked, will be expected to disclose where a particular product is made. Unfortunately in may cases he will have no way of giving the required information. For instance, 'household-name' UK companies are none too keen to admit that some of their goods are being made in low-cost labour countries. Moreover, if a retailer does not know the country of origin, or (even inadvertently) gives the wrong answer, he will be at risk under the Trade Descriptions Act!

Apart from the factors mentioned earlier, which allow a customer to assess the advisability or otherwise of making a purchase, a particularly disturbing aspect of electrical goods (notably electric blankets and appliances) is that safety specifications may fall far short of those imposed on British-made goods and expose the user to potential hazards.

So, together with the British Radio Equipment Manufacturers' Association, the Radio and Television Retailers' Association and the Association of Manufacturers of Domestic Appliances, we beg the Secretary for Trade and Industry to think again. And to would-be purchasers after November 30th we can only say, most emphatically caveat emptor!

W. N. STEVENS-Editor.

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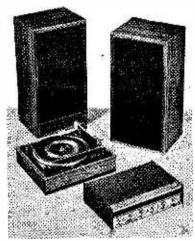
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FREE WALL CHART: Circuit Building Blocks		

NOVEMBER ISSUE WILL BE PUBLISHED ON OCTOBER 8th

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NEWS ... NEWS ... NEWS ...

Viscount value



The Viscount III Audio System is the first of a range of audio equipment marketed in the UK by R+TV Components (Acton) Ltd. under the Viscount brand name.

The complete system includes a choice of two amplifiers, a Garrard SP25 Mk III turntable and two Duo type II or type III speakers.

Amplifier RT100 is for ceramic cartridges and model RT101 which has an additional input stage accepts magnetic types. Both amplifiers incorporate f.e.t.'s giving excellent signal to noise ratios.

Prices are: RT100, £17; RT101, £22; Garrard SP25 with ceramic cartridge, £21 or with magnetic cartridge, £23. Duo type II speakers cost £14 the pair and Duo type III, £32.

Savings can be made by buying complete systems.

Impex cases

In last month's issue we stated the following about Impex instrument cases:

"At no additional cost, Impex cases can be supplied with all or some of these specifications: alternative colours of blue or bronze; no hood; no ventilating slots or back/front anodised front/rear panels".

What we should have mentioned is that these additions can only be supplied free of charge for quantities over 24.

NHK's anniversary

June 1 saw the anniversary of the Radio Japan Service. Thirty-six years ago NHK commenced this service in commemoration of the 10th anniversary of radio broadcasting in Japan and the attainment of 2 million licensed subscribers.

From June 1934 to 1945, NHK broadcast its domestic programmes to Korea, Taiwan and Manchuria, then Japan's outlying territories, utilising telephone channels of the International Telephone Company. Shortwave broadcasts were received in various parts other than the areas directed, and reception reports

began to come to NHK from Japanese residing abroad and from large numbers of overseas people. In those days, Japanese living in various parts of the world totalled nearly two million. These people raised their voices requesting the establishment of an overseas broadcasting service. In Japan, too, commencement of an overseas broadcasting service was also strongly desired, since this country had been isolated after her withdrawal from the League of Nations and was in need of reinforcing its public relations abroad. At first the service was known as Radio Tokyo (Radio Japan was first used in 1952).

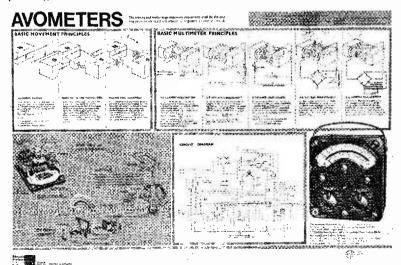
Avo wall chart

Avo Limited (Thorn Group) announce a new wall chart featuring the latest Mark 4 version of the Avometer Model 8. The design and layout is such that it will prove of great interest not only to Avometer users but also to technical schools and general education establishments as a teaching aid.

Information is presented in a clear, logical manner and the theoretical and practical information is both concise and easy to understand. Magnetic theory is developed pictorially to outline the operation of the moving coil movement whilst a mathematical treatment is used to derive the various ranges of a multimeter.

Illustrations of the interior of the instrument include an exploded view of the movement to simplify the constructional detail and a circuit diagram provides an introduction to standard symbols and references.

The Avometer chart (size 20×30in., 508×762mm.) is ideal for black-board or wall presentation and is available from Spares Dept., Avo Limited, Avocet House, Dover, at 75p per copy, including post and packing.



NEWS ... NEWS ... NEWS ...

Windermere VHF

The v.h.f. radio services from the Windermere relay station started on August 7. Radio 2 is on 88.6MHz, Radio 3 on 90.8MHz and Radio 4 on 93.0MHz, all with horizontal polarization.

The v.h.f. radio transmissions from this new relay station, which is situated on Claife Heights, above the western shore of Lake Windermere, will serve Ambleside, Windermere town, Bowness-on-Windermere, and an area around the northern shores of the lake.

Stereo from Rowbridge

The necessary equipment has now been installed at Rowridge to enable the Radio 3 transmitter (90.7MHz) to broadcast stereophony. This is being brought into use immediately, so that listeners can hear the Promenade Concerts in stereo-all except four of the concerts will be broadcast stereophonically in Radio 3. The stereo transmissions from Rowridge will initially be on an engineering test basis; every effort will be made to maintain them, but there may be occasions when it will be necessary to revert to monophonic transmissions. The date for the start of the full stereophonic service will be announced later.

Miniature Electric Tools

A range of battery-operated miniature tools has been introduced by Expo (Drills) Ltd., 62 Neal Street, London, W.C.2.

Two basic models are available. The 'Reliant' is designed for lighter work such as model making, and has a full load current of 1.5A. Its rated torque is 1.38oz. in. (100 gm.cm.). For jobs requiring a more powerful tool and for professional applications the 'Titan Super' is rated at 3.5A on full load. It has a rated torque of 350cmp operating at 4000-9000rpm. An extremely large range of accessories is also available for both models. Various collets and accessories are supplied with the different models, which range in price from £3 to £5.50.

Lectures

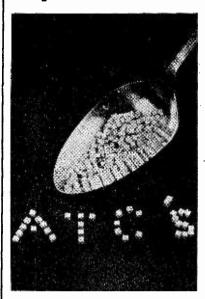
THE Polytechnic of North London Department of Electronic and Communications Engineering is running a course on Audio and Acoustic Measurements on Thursdays 6.30-8.30 p.m. 1971/2. 1971: Oct., 28, Parameters and Standards of Measurement R. N. Baldock, B.Sc., F.B.K.S.T.S. Nov. 4, Measurement of Microphone Performance, H. D. Harwood, B.Sc., B.B.C. Research: 11. Electrical and Mechanical Measurements in Tape Recording, R. L. West, B.Sc., C.Eng., F.I.E.R.E.; 18, Measurements in Professional Recording, Angus McKensie; 25, Electrical and Mechanical Measurements on Disc, Pickups and Accelerometers, S. Kelly, C.Eng., F.I.E.R.E. Dec. 2, Measurement of Loudspeaker Performance I, R. L. West; 9, Measurement of Loudspeaker Performance II, R. C. Driscoll, M.Sc.: 16. Measurement of Studio Characteristics, R. C. Driscoll, M.Sc. 1972: Jan. 13, Measurement of Noise Transmission, J. Moir C.Eng., F.I.E.E.; 20, Assessment of Noise Nuisance, J. Moir, C.Eng., F.I.E.E.; 27, Visit to A.I.R.O. Laboratories, Hemel Hempstead, G. Berry, B.Sc., C.Eng., M.I.Mech.E. and Assocs.

Grand Junk Sale

The Star Short Wave Club, G3ZWA, is holding a "Grand Junk Sale" in aid of Radio Amateur Invalid Bedfast Club funds. It is to be held at the New Inn Hotel, Bramley Town Street, Bramley, Leeds 13. The date is Wednesday September 15th and time 7.30.

If any readers have any junk, the Star Club will collect within a radius of 25 miles from Leeds. Please contact Mr. T. Leeman, G8BUU, 115 Asket Drive, Seacroft, Leeds, LS14 1HX.

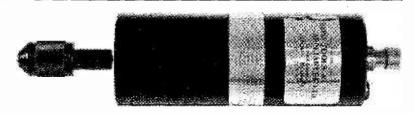
Chip "C's"



Featuring a very low insertion loss, a very high Q and high capacitance-to-volume ratios, a new range of ceramic and porcelain chip capacitors from Guest International Limited is ideal for u.h.f. and microwave applications. Apart from the basic pellet and chip, there is a wide choice of lead variations: stripline, axial and radial wire, axial and radial ribbon.

ATC 100 series are porcelain chips with an insertion loss of 0.03dB, a Q of over 10,000 and a capacitance range of 0.1 to 1000pF. In a case size of only 0.110in.square, the working voltage is up to 500V d.c.

ATC 200 series are ceramic chips with an insertion loss of 0.05dB. The capacitance range is 10pF to 1.2μ F and chips are available in case sizes 0.050, 0.110 and 0.250in square, depending on the value and the working voltage.





So many audio sources are now in common use that an article describing a specific design would meet the needs of only a few. This problem is overcome here by giving circuits for virtually all sources and care in design has made possible a large number of permutations so that a custom built audio mixer is possible even for those readers unable to design their own. The clever constructional method employed also allows for rapid modification to meet any mixing need.

STUDY of the four block diagrams in Fig. 1 will show how the various signal preamplifiers, the tone control unit and line amplifier etc. can be used to build audio signal mixing systems to cater for tape recording, public address, discotheque and even Hi-Fi applications. The combinations given are but a few of the permutations which can be extended for stereo use simply by duplication.

Each signal preamp has the same output level, 100mV, and will match directly into either the tone control unit or the line amplifier, both of which are designed to accept a signal level of 100mV. The tone control has unity gain so this provides 100mV output to match into the line amplifier. Any group of signal preamplifiers can therefore be coupled directly to the line amplifier or via the tone control unit and then to the line amplifier. This allows any of the signal preamplifiers to be used with or without the tone control unit.

The line amplifier has a nominal output of 1V r.m.s. at a fairly low impedance and can be used to drive a power amplifier module requiring between 500mV to 1V input, or it can be coupled to a tape recorder or radio input, of to a public address or music amplifier with inputs rated for 500mV to 1V. Groups of the preamplifiers and line amplifier can be run from batteries or from an 18V mains power supply.

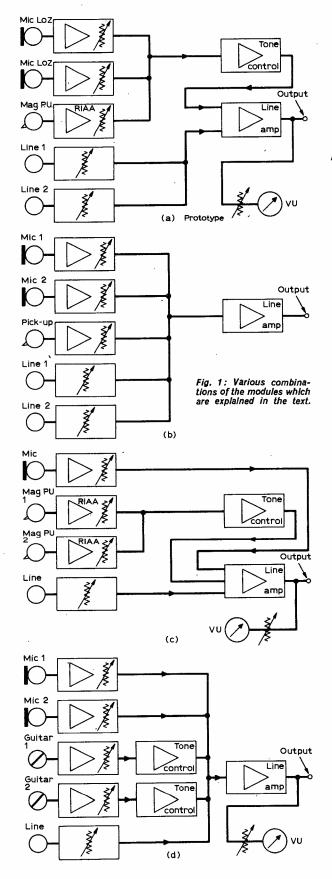
A fairly simple arrangement is shown in Fig. 1a and is intended for use as a general purpose signal mixer as shown on the front cover. This would be suitable for tape recording, as it has two microphone inputs, one input for low impedance pick-up (with R.I.A.A. correction), and two line level (200mV to over 1V) inputs. The tone control operates for the mic. and pickup inputs, whilst the line inputs go directly to the line amplifier. The system also includes a VU meter with variable control so that the meter can be set to read with the level meter on a tape recorder.

The combination shown in Fig. 1b has the same input channels but no tone control unit. The arrangement in Fig. 1c might be suitable for a small discotheque with one mic. input, two pickup inputs both via a tone control unit and one line input, suitable for the output from a tape recorder. For music application, Fig. 1d provides two mic. and two guitar inputs with separate tone control units for each guitar input. In each case the VU meter is optional but can be used as an overload indicator as will be shown later.

PERFORMANCE SPECIFICATIONS

Seven preamplifiers will be described, each catering for specific signal sources as follows:—

- LOW IMPEDANCE MICS. (25 to 600Ω). Direct input, no transformer required. Input sensitivity 50 V. Output 100mV. Frequency response ±1dB.
 30 to 30,000Hz. Signal-to-noise -50dB. (Code: Mic. Lo-Z).
- FOR MICROPHONES WITH BUILT-IN TRANS-FORMERS. Medium to high impedance. Input sensitivity 1mV. Output 100mV. Frequency response ±1dB 20 to 30,000Hz. Signal to noise -60dB (Code: Mic. Hi-ZT).
- FOR CRYSTAL MICS. High impendance. Input sensitivity 2mV. Output 100mV. Frequency response ±1dB 30 to 12,000Hz. Signal-to-noise -50dB. (Code: Mic. Hi-ZX).
- MAGNETIC PICKUP Impedance 56kΩ. Input sensitivity 5mV. Output 100mV. Frequency response to R.I.A.A. as shown in Fig. 2. Signal-tonoise -60dB. (Code: PU Mag.)



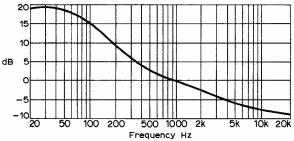


Fig. 2: The RIAA response curve incorporated in the magnetic pickup module.

- CERAMIC PICKUP. Input impedance 2MΩ. Input sensitivity 60mV. Output 100mV. Frequency response approximately 20 to 20,000Hz ±1dB. Equalisation automatic due to high input impedance. Signal-to-noise -50dB. (Code: PU Cer.)
- CRYSTAL PICKUP, Input impedance about 1MΩ. Input sensitivity 850mV. Output 100mV. Frequency response approximately 20 to 20,000Hz ±1dB. Equalisation automatic due to high input impedance. Signal-to-noise -60dB. (Code: PU Xtal.)
- 7. GUITAR PREAMP. Input impedance medium, 10 to $100 \mathrm{k}\Omega$. Input sensitivity variable. 50 to $100 \mathrm{mV}$. Output $100 \mathrm{mV}$. Frequency response $\pm 1 \mathrm{dB}$ 20 to $30,000 \mathrm{Hz}$. Signal-to-noise -50 to $-60 \mathrm{dB}$. (Code: G.)

The tone control unit will take the output from any of the preamps and provides 100mV output so that it can be coupled directly to the line amplifier. Its performance is as follows:— With tone controls neutral the flat response is 10 to 80,000Hz ±1dB. The response of the bass and treble controls is shown in Fig. 3 providing 15dB lift and 12dB cut at 40Hz and 15dB lift to 12dB cut at 20,000Hz The signal to noise performance is better than -90dB. (Code: **TC**).

The line amplifier unit will accept the output from any of the preamps or the tone control unit and therefore has an input sensitivity of 100mV. The output signal level available is 1V r.m.s at an impedance of approximately 6000. The frequency response is flat from 10 to nearly 100,000Hz and the signal-to-noise performance —80dB. The output can be fed simultaneously to one or more tape recorders and/or amplifiers via series output resistors as will be shown. The overall gain of the line amplifier is adjustable i.e., the input sensitivity can be increased. (Code: LA.)

The line inputs do not require separate preamplifiers as these are arranged in passive network coupling directly to the line amplifier. The input sensitivity will be 200 mV to over 1 V r.m.s. at approximately $10 \text{k}\Omega$ impedance. (Code: L input.)

CONSTRUCTION

The construction of each of the preamplifier modules is such that the circuit board is mounted on a small screen this, in turn, being secured to the fixing bush of the gain control (VR1 in each circuit). The complete preamp can therefore be secured to a front panel by the lock nut of the gain control. The tone control unit is similar except that there are two potentiometers and both are mounted on the small screen and then secured to the panel. The line amplifier can be simply mounted on a piece of aluminium angle $(^3_8 \times ^3_8 \text{in})$ or on stand-off pillars as convenient. It need not be fixed to a panel but must

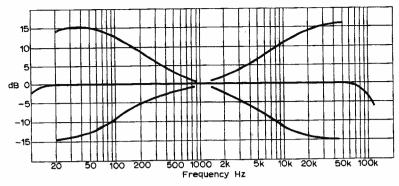


Fig. 3: Response curves of the tone control module.

be fairly well clear of mains power supply components. The input sockets for all inputs must be insulated from the front panel and earthed only at the common earthing point on the preamplifier board. If the mixing unit is to be mains powered, the components for the power supply must be kept as far away from the amplifier boards as possible and preferably with a screen around them.

The seven preamplifier boards are exactly the same and details for construction are given in Fig. 4. In each case the gain control (VR1) is fitted as shown

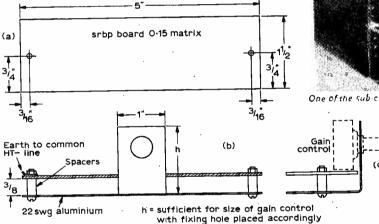
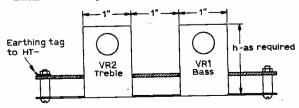


Fig. 4: Details of the sub-chassis used for each module.

in Fig. 4c. The construction of the tone control board is similar except that two controls have to be fitted so the screen has two flaps as shown in Fig. 5. The height of each flap and the position of the fixing hole will depend on the physical size of the potentiometers used.

CIRCUITS AND ASSEMBLY OF THE MODULES

Each of the amplifier modules and the tone control



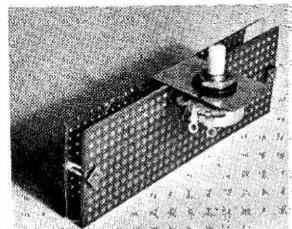
Controls mounted as in Fig. 4 (c)

Fig. 5: Modified sub-chassis for the tone control module.

can be tested separately by operating them from an 18V battery (two 9V units) and by checking each one with an audio signal generator and audio voltmeter or by connecting an appropriate input signal source and feeding the output to an amplifier.

LINE AMPLIFIER

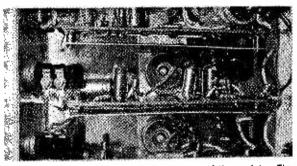
The best order of construction is to build the line amplifier (code: LA) first so that the preamps can be tested with it as each is completed. The circuit is given in Fig. 6 and the board layout in Fig. 7. As



One of the sub-chassis before the components are fitted; dimensions are given in Fig. 4.

with all the circuits except the tone control unit only one type of transistor is used—the BC109 silicon NPN—which is readily available and inexpensive. The line amplifier employs three of these beginning with Tr1 as an emitter follower input stage with Tr2 as the amplifier.

Negative feedback between Tr2 and Tr1 via the pre-set PR1 ensures a linear response and some adjustment over total gain. The pre-set should be adjusted so that with a 100mV input signal the output at C6 is exactly IV. If a signal generator and audio voltmeter are unavailable, set PR1 to midway position. The output transistor Tr3 operates as an emitter follower and provides a low impedance



An internal view of the prototype showing one of the modules. The input jack is on the left with the common wiring on the right.

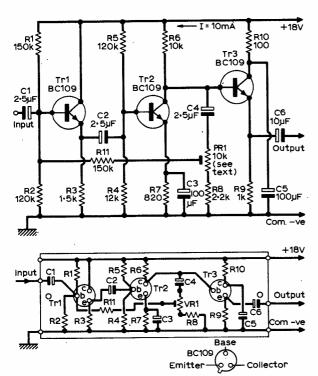
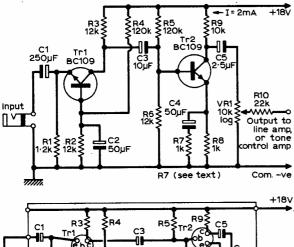


Fig. 6: (above) and Fig. 7 (below): The circuit and layout of the line amplifier module.

output. Details for feeding more than one power amplifier and/or tape recorder from the line amplifier output will be given later.

MICROPHONE PREAMP (Mic Lo-Z)

The circuit is given in Fig. 8 and the board layout



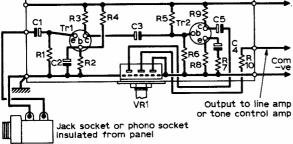
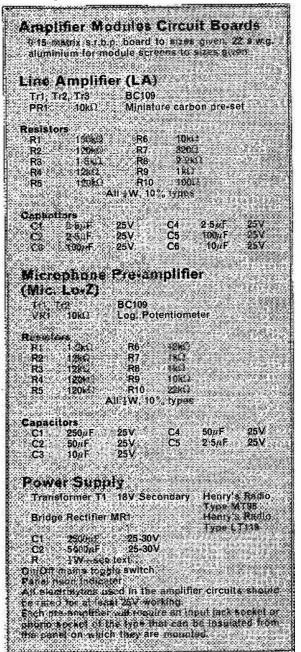


Fig. 8 (above) and Fig. 9 (below): The circuit and layout of the low impedance microphone preamplifier module.

★ components list



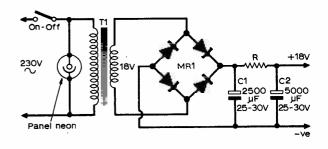
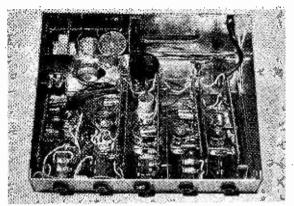


Fig. 10: Suggested circuit for a mains power supply to power the mixer.

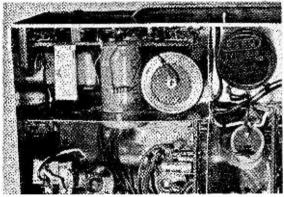
in Fig. 9. The input transistor Tr1 operates as a grounded base amplifier and provides a very low impedance input. Tr2 is a signal amplifier with negative feedback introduced by R7 in series with the emitter by-pass capacitor C4. This resistor can be reduced a little in value to obtain more gain for very low output microphones, although it should not be less than 680Ω otherwise the noise level will rise excessively. As with all the preamplifiers the output is taken to the gain control VR1 and thence via the series resistor (R10) which becomes part of the passive mixing network common to all the preamps.

MAINS POWER SUPPLY

The current consumption of the preamplifiers at



Internal view of the completed prototype.



The power supply should be properly screened as shown.

18V varies between about 2 and 5mA. The line amplifier takes 10mA and the tone control unit 5mA. Four preamplifiers, a line amplifier and say one tone control unit, would require about 25mA which could be drawn from an 18V battery with reasonable economy. Two PP9 batteries would be ideal. Larger mixing units, particularly if built for stereo use and requiring duplication of preamps etc. are best operated from a mains power supply. One that will provide well over 100mA is given in Fig. 10 (see components list for transformer and rectifier details). The resistor R will be around $1k\Omega$ for low current drain and of lower value for higher current drain. With the mixing circuits running, adjust R to obtain 18V across C2.

Part 2 next month, will deal with the rest of the preamplifier modules.

TO BE CONTINUED

edium ave Column

NEW 1000kW station on 1457kHz located near Tirana in Albania is being heard in this country as a background to BBC Radio 4. (South and West). It relays Radio Peking and has been heard at 2100hrs GMT with the identification 'Govarit Pekin' followed by programmes in Russian. An increasing number of superpower broadcasters are appearing on the medium waves. Urumchi in Western China (2000kW) is audible all evening on 1525kHz; listen between Prague 1520kHz and Vatican Radio 1529kHz for Russian speech plus jamming. Libva has two megawatters—El Beida on 1124kHz and Tripoli on 1250kHz. El Beida is the stronger of the two; it is heard with EAJ15 Barcelona on the same frequency throughout the evening. Tripoli suffers from interference from Dublin/Cork until 2245hrs when the latter close down, leaving the channel clear. Further east, Riyadh (1200kW) in Saudi Arabia is on 587kHz and has been logged before midnight with Arabic programming.

Calcutta 1130kHz (1000kW) was heard frequently last winter before sign-off at 1700hrs. It is now on extended schedule and has been several times during the summer at 2230hrs GMT, as it changes from Chinese to Indian programming. There are three Voice of America stations in the Far East, each with 1000kW, all of which have been heard in the U.K. Bangkok, Thailand 1580kHz was logged at the end of September last year at 2200hrs GMT while Poro 1140kHz in the Phillipines and Okinawa 1178kHz in the Ryuku Islands are both heard in winter before

sign-off at 1700hrs.

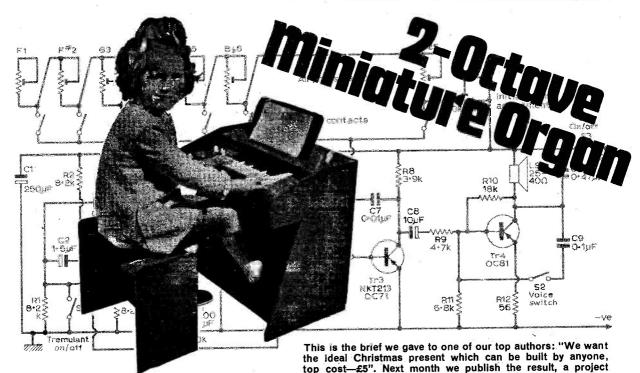
'Book of the World 1971' is the title of an unusual paperback now on sale in the bookshops. Sections deal with 181 different countries, each with a map plus details of the type of government, languages spoken, religions and currency. As well as providing the DXer with a valuable aid to station identification, this book gives background information on countries

logged and adds interest to the hobby.

Broadcasts in English on the medium waves can be heard from a number of unexpected places during the late evening. Timisoara, Rumania on 755kHz has a programme in English, French and German at 2230hrs GMT, which should not be confused with the Voice of the West, in Portugal, also on 755, which starts its English programmes at 2245hrs. The Voice of America in Rhodes is 1259kHz is heard regularly in English between 2100hrs and midnight, with some interference from Wroclaw in Poland. Rhodes is one of the Dodecanese Islands situated near the coast of Turkey, which are usually counted as a separate country by DXers. The VOA is also heard in English from Thessalonika, Greece 791kHz between 2100hrs and 2130hrs. Enugu, Nigeria 1320kHz, although only 10kW, is a weak but consistent signal at 2300hrs GMT when it carries it's final news in English before closing down. ELBC Monrovia, Liberia on 629kHz has English programming and is usually logged between 2330hrs and sign-off at 0045hrs.

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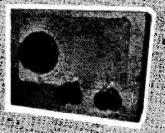
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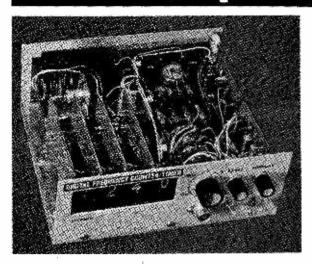
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DIGITAL FREQUENCY COUNTER/TIMER



PART 2 (continued from the September issue)

This negative going step is passed by C5 to Tr7 base which moves by -5 volts thus cutting off Tr7 and allowing the collector of Tr7 to rise to +5V.

The current through VR2 and R14 discharges C5 so that the negative voltage on Tr7 base rises exponentially until Tr7 conducts. The rate is determined by the setting of VR2, the DISPLAY control.

When Tr7 conducts, its collector voltage starts to fall, turning off Tr6 and thus increasing the base current to Tr7 and producing a fast negative going voltage at Tr7 collector. This voltage step is differen-



J. THORNTON-LAWRENCE GW3JGA

NOTE

Several wiring connections were omitted from figure 6 in the September issue. Pin 10, on IC's 3, 4, 5, 6 and 7, should be shown connected to the OV rail. Pins 1 and 12 on IC's 9, 11, 13 and 15, should be shown joined.

On IC1, pin 4 is the +5V supply rail and pin 11 is OV supply rail. On IC2, pin 14 is +5V supply rail and pin 7 OV supply rail. These latter connections were omitted from figure 6 for clarity.

tiated by C7, R16 and R17 and inverted by IC2d to give a positive pulse of 200 µS duration.

The positive going edge of this pulse resets the counter to zero and the negative going edge sets the JK flip-flop ICla and causes the cycle of operation to be repeated. The time for which the reading is displayed is controlled by the DISPLAY control and this is adjustable from about 0.5 second to 2 seconds.

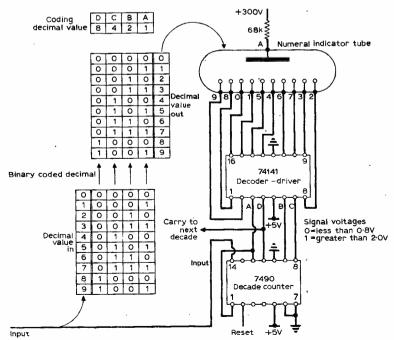


Fig. 9: Table showing the operation of the decade counter and decoder-driver.

Counter and Display

The counter and display has four decades, each decade operates in the same way. The output of one decade is linked to the input of the next higher decade to perform the 'carry' operation and a reset connection is common to all decades.

The input to the counter is a series of pulses. The counting has to be carried out in binary form and then decoded into decimal form for display.

This is done using a decade counter I.C. type SN7490N, which consists of 4 master-slave flip-flops forming $a \div 2$ counter and $a \div 5$ counter.

The input signal goes to the input of the +2 counter (pin 14) and the output of this (pin 12) is linked externally to the input of the $\div 5$ counter

(pin 1).

The outputs of the four flip-flops are called A, B, C, D and have a decimal value of 1, 2, 4, 8 respectively. The table showing the operation of the decade counter is included in Fig. 9.

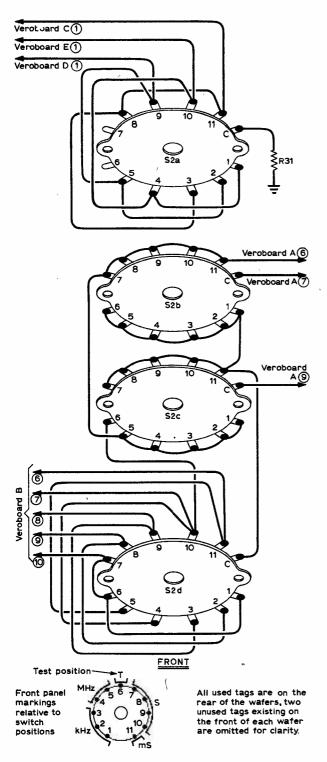


Fig. 10: Exploded view of the function switch S2, viewed from the front.

The counter can be reset to zero at any time by the application of a positive pulse to pin 2.

The binary coded decimal output from the counter, A, B, C, D is connected directly to the inputs of the decoder/numeral tube driver I.C. type SN74141N. Here the inputs are decoded back into decimal form

for driving the numeral tube (Hivac Type XN13).

The numeral tube contains neon gas at low pressure and has an anode and 10 wire cathodes in the form of numerals, 0-9.

With the anode connected to HT+ through a current limiting resistor, any cathode which is connected to 0V will be surrounded by a glow showing up the shape of the number.

The outputs of the decoder-driver are collector circuits of high voltage transistors which, when made to conduct, effectively connect the appropriate cathode to 0V. For example, suppose the input to the decade counter receives 5 pulses or cycles of input signal. The output will be D low, C high, B low, A high, representating a binary number of 0101. This number is decoded in the decoder/driver to figure 5 and the figure 5 output (pin 14) becomes switched to 0V causing the figure 5 to be displayed in the numeral tube.

Each counter, decoder/driver and tube is built on a Veroboard panel and the inter connection of these panels is shown in the main diagram Fig. 6.

The decimal indication is provided by miniature wire ended neon indicator lamps mounted between two numeral tubes and these are switched by the function switch S2a.

Function Switch

The function switch S2 is made up from Radiospares Miniature 'Maka Switch' parts and consists of a shafting unit and 4 break-before-make 1 pole 12 way wafers. The wafers are spaced by 2 spacers on each side. Only 11 positions are used and the stop on the indexing mechanism is adjusted accordingly.

The wafers are lettered a, b, c and d, starting at the rear and working towards the indexing mechanism. Wafer 'a' controls the decimal point neon lamps, wafer 'b' the clock period input to the gate, wafer 'c' the signal input to the gate and wafer 'd' the output from the clock divider.

In the frequency positions 1 to 5, the output from the input stage is fed to the 'Signal Input' of the gate and the appropriate clock signal is fed to the 'Clock Period Input' of the gate. In the time positions, 7 to 11 of the switch, the output from the input stage is fed to the 'Clock Period Input' of the gate and the appropriate clock signal is fed to the 'Signal Input' of the gate.

A test position is provided in position 6 where the 1kHz clock output is counted for a period of 1 second to give a reading of 1,000. The test position only serves to check that the gate, counter and display are functioning, it does not provide a check of the crystal oscillator frequency.

The wiring and connections to the switch are shown in Fig. 10. The method of assembly is as follows:

- 1. On the shafting unit, check that the lengths of studding are screwed fully into the indexing mechanism plate and remove the end nuts and washers.
- 2. Rotate the switch spindle to the fully anti-clockwise position (as viewed from the front). The flats on the wafer end of the spindle should be in line with the studding. This is position "1".
- 3. The switch should now be held so that the lengths

of studding are at each side and the control knob end of the spindle is towards you, as shown in Fig. 10.

- 4. Take a 1 pole 12-way wafer and hold this with the side having the majority of tags, away from you and with the two tags that are facing you positioned in the top right hand segment. Rotate the rotor section until the common or rotor contact is making contact with tag 1. The slot in the rotor should now be horizontal.
- 5. Assemble the wafer on to the shafting assembly. keeping the side with the majority of tags facing away from the indexing mechanism. This wafer is now wafer "d" in Fig. 10.
- 6. Fit 2 spacers on each piece of studding.
- 7. Fit another wafer as described in operation 4 and 5 (wafer "c").
- 8. Repeat operations 6 and 7 until four wafers have been fitted (wafers "b" and "a").
- 9. Fit washers and fixing nuts to the studding.
- 10. The rotation of the switch should now be checked to see that all parts operate smoothly.
- 11. The rotatable stop plate, which is located on the face of the indexing mechanism, should now be set to give 11 positions, working clockwise and commencing from the maximum anti-clockwise position. This is the position at which all the wafers have been set. The stop plate may be held in place temporarily by the mounting nut, ready for assembly to the front panel.
- 12. The wiring of the switch may now be carried out as shown in Fig. 10. 22 s.w.g. tinned copper wire is used for wiring from tag to tag, sleeving being necessary only when crossing between wafers.
- 13. The correctness of the wiring can be checked by using an ohm or continuity meter and measuring between the appropriate tags whilst the switch is rotated, e.g. S2b/C is connected to S2d/C in positions 1. 2. 3. 4. 5 and 6.

Power Supply

-5V

The power supply consists of a Radiospares Midget Mains transformer which with its associated rectifiers smoothing and regulators provides

> +300V at 10mA Unregulated Regulated +5Vat 350mA Regulated

at 5mA

The output from the 250 volt secondary winding is rectified by D7 and smoothed by C12 to give about +300V output. The negative return is through R24 and D9 to 0V. The negative supply derived in this way is smoothed by C13 and regulated by the zener diode D9 to give -5V. The I.C. logic requires about 350mA at +5V and this is provided by the 6.3 volt filament winding on the transformer.

The 6.3 volt output is full wave bridge rectified to give about +7.5 volt across the reservoir capacitor C14. This feeds the zener diode D6 and the power emitter follower Tr11 to give the required 5V output.

Veroboard Panels

The layout drawings of the 0.1×0.1 in matrix Veroboard show the copper strip side of the panel the components on the reverse side being through" the copper strips.

Each copper strip is given a letter code and each

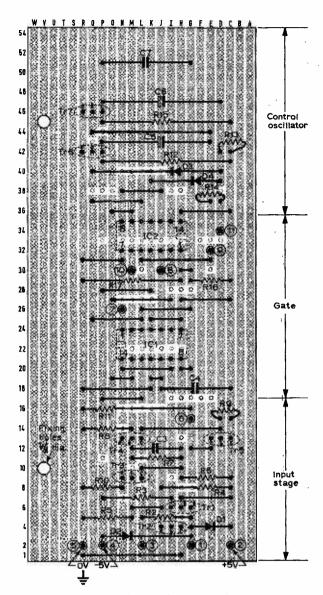


Fig. 11. Veroboard panel A. Input stage, gate and control oscillator. Viewed from copper side. Note correct orientation of integrated circuits when fitting. The integrated circuits and their holders are indented at one end to assist orientation.

horizontal row of holes, a number.

All solid horizontal lines represent wire links which connect one strip to another.

The orientation of integrated circuits and transistors should be noted carefully, some transistors may have unusual lead positions and these may have to be crossed over for the connections to be correct. Assembly of the components, etc., must be carried out in a systematic manner or errors will result. A suggested method is as follows:

- 1. Cut the Veroboard to the correct dimensions using a fine toothed saw and working from the copper side. Apart from checking the physical dimensions, also check that the correct number of holes appear in each direction.
- 2. Fix strips of masking tape on all four edges of the Veroboard panels. Also fix masking tape to the same edges on the reverse (copper) side of the board.

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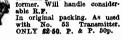
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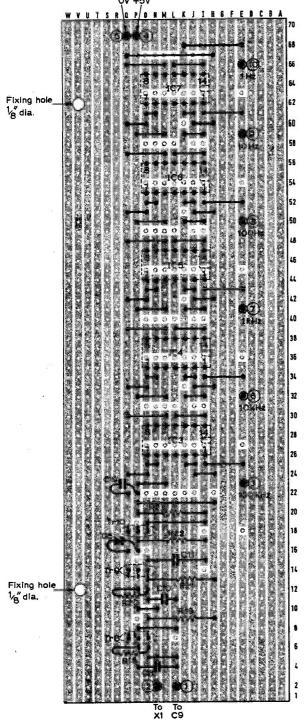


Fig. 12. Veroboard panel B. Crystal clock and divider. Viewed from copper side

- 3. By writing on the masking tape, letter all the vertical strips and number all the horizontal rows of holes on both sides of the board.
- 4. The breaks in the copper may be made using the appropriate Vero tool, part No. 2022 or by using a rain drill held in a pin vice. Each copper break on the layout drawing should be identified by its co-

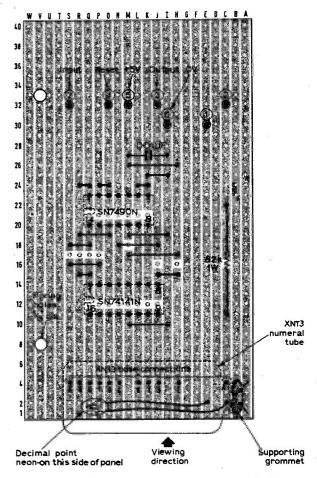


Fig. 13: Counter and display panels C. D. E. F. Panel F. does not require a neon "decimal".

ordinates, e.g. D38 and the copper strip removed by carefully rotating the tool or drill by hand using the identified hole as a guide.

Experience has shown that just counting the position of a hole is not reliable and errors are very likely to result.

5. Wire links are formed from pre-stretched 22 s.w.g. tinned copper wire. A length of about 3ft of 22 s.w.g. TC wire is stretched slightly by using a pair of pliers at each end, until it becomes stiff. All wire links are fitted on the components side.

A length, in excess of the length of the link required, is cut off and two right angled bends are made in the appropriate places to form the required "U" shaped link. A pair of pointed nosed pliers is useful for this and the span of the link can be checked by holding it against the panel in the appropriate place. When the link is inserted, the end wires may be bent outwards slightly to hold the link in place while it is soldered.

6. Components should be fixed starting at one end of the panel and the position and value checked before soldering.

7. The soldering should be carried out using a soldering iron with a small chisel shaped bit, no wider than $\frac{1}{16}$ in. The cored solder used should be no larger than 20 s.w.g.

Check each soldered joint using a magnifying glass as it takes only a very small piece of swarf

or solder to bridge the 0.025in gap between the copper strips.

8. Connections to the Veroboard panel are made to pins inserted in the appropriate places and soldered to the copper strip. These pins are Vero type TP 11032 or similar.

An insertion tool for this pin is Vero part No. IT 2151.

9. Tick off each break, link, component and pin on the drawings as they are fitted, this will identify any items which otherwise might be missed.

Semiconductors

The 3702 and 3704 transistors are specified in the components list as being the '2N' types, these are the most popular version and are readily available. The 3702 and 3704 transistors are also available

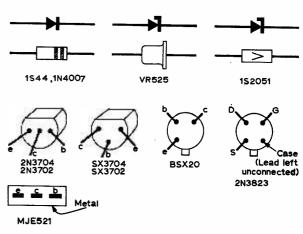


Fig. 14 (left) Cabinet and chassis construction.

in an 'SX' version. The lead connections for these types are all different from each other and are shown in the diagram above.

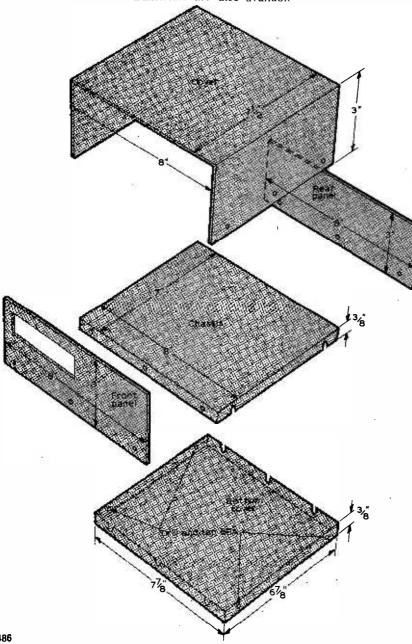
It should be noted that on the Veroboard panels, the connections run e.b.c. and when the '2N' version is used. the collector and base leads will have to be crossed over to line up with the appropriate holes in the Veroboard panel.

Sockets for Integrated *Circuits*

Sockets are used for all the integrated circuits and although this increases the overall cost and is not essential for the operation of the instrument, it does provide the facility for disconnecting or changing an I.C. should a fault condition exist, also it does allow various parts of the instrument to be tested without incurring the expense of buying all the I.C's at once, 16-way sockets are required for I.C.8, 10, 12 and 14 and 14-way sockets for all the others.

The instrument is built using a simple aluminium chassis, panel and cover type of construction, as shown in Fig. 14. The finished chassis size is $8 \times 7 \times 3_8$ in deep and has a front and rear panel, each size 8×3in, fastened to the front and rear edges with 6BA screws and nuts.

(To be continued)



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2N696	17p	2N2924	20p	AC107	46p	BC153	19p	BFY50	231
2N697	18p	2N2925	22p	AC126	20p	BC154	20p	BFY51	200
2N706	12p	2N2926	11p	AC127	20p	BC157	12p	BFY52	231
2N930	29p	2N3053	27p	AC128	20p	BC158	11p	B8X20	161
2N1131	29p	2N3055	60p.	AC153K	22p	BC159	12p	C407	17
2N132	29p	2N3702	13p	AC176	16p	BC167	11p	MC140	251
2N1302	19p	2N3703	13p	ACY20	20p	BC168	10p	MP86531	351
2N1302	19p	2N3704	13p	ACY22	16p	BC169	11p	MPS6534	801
2N1304	26p	2N3705	13p	AD140	63p	BC177	14p	NKT211	251
2N1304	26p	2N3706	13p	AD142	50p	BC178	13p	NKT212	251
2N1306	33p	2N3707	13p	AD149	58p	BC179	14p	NKT214	23
2N1307	83p	2N3708	10p	AD161	33p	BC182L	11p	NKT274	181
2N1308	36p	2N3709	llp	AD162	36p	BC183L	10p	NKT403	65
2N1309	36p	2N3710	13p	AF114	24p	BC184L	11p	NKT405	79
2N1613	23p	2N3711	13p	AF115	24p	BC212L	16p	OC71	38
	28p	2N3819	23p	AF117	22p	BC213L	16p	OC81	25
2N1711 2N1893	54p	2N3904	35p	AF124	24p	BC214L	16p	OC81D	25
2N1093 2N2147	95p	2N3906	35p	AF127	22p	BCY70	18p	ZTX300	14
2N2218	34p	2N4058	13p	AF139	33p	BCY71	33p	ZTX301	16
2N2218A	44p	2N4059	100	AF239	36p	BCY72	15p	ZTX 302	22
2N2218A 2N2219	38p	2N4060	11p	ASY26	27p	BF115	23p	ZTX303	22
2N2219A	53p	2N4061	110	ASY28	27p	BF167	18p	ZTX304	27
2N2219A 2N2270	62p	2N4062	12p	BC107	12p	BF173	19p	ZTX500	18
2N2369A	19p	2N4124	18p	BC108	110	BF194	14p	ZTX 501	21 25 22
2N2369A 2N2483	35p	2N4126	27p	BC109	12p	BF195	15p	ZTX502	25
		2N4284	15p	BC125	15p	BFX29	31p	ZTX503	22
2N2484	42p	2N4286	15p	BC126	22p	BFX84	25p	ZTX504	52
2N2646	47p		15p	BC147	10p	BFX85	32p		
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Audio Audio Audio		 2-pole 3-pole 4-pole 5-pole 180 deg.	plug 12p 13p 14p 15p	30cket 10p 10p 12p 12p	
		5-pole 240 deg.	15p	12p	
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Code	Power	Tolerance	Range	
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č	1/8W	5%	4 7 Ω-470K Ω	
Ċ	1/4W	10%	4.7 Ω -10M Ω	
C	1/2W	5%	4·7 Ω-10M Ω	
Ċ	1W	10%	4.7 Ω -10M Ω	
MO	1/2W	2%	$10 \Omega - 1M \Omega$	
ww	1W	$10\% \pm 1/20 \Omega$	$0.22 \Omega - 3.9 \Omega$	
ww	3W	5%	12 Ω-10Κ Ω	
ww	7W	5%	$12 \Omega - 10 K \Omega$	
Codes:	C = carbon	film high stabil	ity low noise	

MO = metal oxide Electrosii Thō ultra low noise
WW = wire wound Plessey.

Values: E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 62, 75, 91 and their decades.

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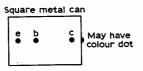
THE technical press offer so many transistor projects that many of us have bought and acquired a fair stock of transistors gathered from bargain packs or labouriously removed them from surplus computer boards. We then go through them only to find that many are un-marked, indecipherable, or have some queer markings not contained in any lists. The manufacturers are reluctant to supply data because they have been made to "customer requirements". A few hours spent in sorting these into NPN or PNP and grading them into four categories will save a lot of time, patience and money on the next project development. The methods of grading are only approximate and there are only four grades: -

- 1. RF types.
- 2. Low gain types.
- 3. High gain types.
- 4. All others which do not pass the tests described.

PRELIMINARY TESTING

First do a thorough visual examination and sort the transistors into known and unknown. The knowns, however, should be put through the tests since they may be blown or below spec. If your collection contains transistors removed from computer boards, the leads will be short. Fortunately when these boards are assembled, small plastic

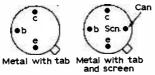
Fig. 1: Some transistor lead connections. These are only the most common arrangements and variations do exist.













spacers are fitted, in most cases to comply with manufacturer's recommendations as to bending and heat dissipation. The makers also usually bend over about 18in on the solder side with the result that the lead lengths are in a safe condition. Any that are too short should be thrown away. Figure 1 shows the bases of seven categories into which these recovered units fall. Some may have a coloured dot, usually red or white, indicating the collector lead, also some are internally connected collector-to-can. Do not be

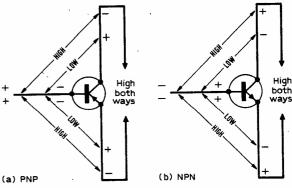


Fig. 2: Resistance readings found on a good transistor—see text for details.

fooled however by coloured sleeving on the leads; not all manufacturers use the same code. Of course a really good transistor checker can make things very easy but this will cost a good deal of money.

May have

Metal or glass

IDENTIFYING THE LEADS

The first checks, sorting the PNP and NPN types and the leads are done with a multimeter on its low ohms range. A word of warning first about that multimeter. Most meters on the ohms range give positive

volts at the NEGATIVE LEAD and negative at the POSITIVE lead, so check first. Figure 2 illustrates the method of finding out whether the transistor is PNP or NPN. Never be scared of reversing the leads from the meter; with about 1.5V, which is usually available from the meter on low ohms ranges, you won't do any harm. Should you get a low resistance reading both ways between base and either of the other connections you can be fairly certain that there is an internal short. If you get a low reading between the base and one lead and a high at the other it is either bad or you are not on the base. It is probable that you will meet with marginal readings at times, so make tests as follows with reference to Fig. 3.

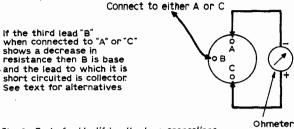


Fig. 3: Tests for identifying the base connections.

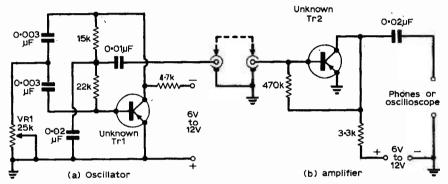


Fig. 4: The circuit used for the gain and amplification tests. Note that for NPN devices the battery polarity must be reversed.

Draw a base diagram of the base of the transistor and mark them 'A', 'B', 'C'; you are now faced with three permutations, of pairs 'AB', 'AC' 'BC'. Now take any pair and connect to the ohmmeter on low ohms range, then connect the other lead to each lead in turn: if the resistance falls the unpaired lead is the base lead and the lead to which the base was shorted is the collector. Check to see if this lead is connected to the metal can, this could confirm that it is the collector. If you do not get these results then check each of the three pairs using the same procedure. If the base cannot be identified for certain and there is high resistance both ways it is most probable that the third lead is base. To check, assume this is base and check with the meter whether it is a PNP or NPN in Fig. 1. You should then have established the base connection; the emitter and collector connections will be identified by reversing the leads in the grading test.

TESTING FOR GAIN ·

To test for gain two simple circuits are required as shown in Fig. 4. Good quality transistor sockets are essential since we are most likely dealing with short leaded transistors and soldering in and out will become tedious. The assembly of the the circuits is an individual choice, though the author used 'S-Dec' and found it ideal. When the test rig is assembled, put in two good known transistors, say an OC44 or OC45 for Tr1 and an OC70 or OC72 for Tr2, connect up the battery and headphones. If you can use a

'scope, all the better. Now adjust VR1 until a tone (or trace) is observed, then start to substitute the unknowns for Tr1. If a tone is present the gain will be more than 30, if no oscillation occurs then it is less than 30. At this point those transistors which have connections which are difficult to identify can be checked and identified by insertion, keeping the assumed base connection and reversing the other two unknown leads in the socket. If no oscillation occurs, keep it for the two further tests. Do not be afraid to reverse leads even at this higher voltage; if it blows then it was probably useless anyway.

AMPLIFICATION TEST

Now replace the original test transistor i.e. the OC44/45 (or one of the good ones known to oscillate) and then one by one test all those which failed the high gain test by substitution in the Tr2 position. The note heard in the headphones will tell you if it will amplify at all. If it is possible to use an oscilloscope, a little more information will be gained on this last test which is for low gain. By means of the trace

shown on the 'scope, you can classify some transistors in the large signal class because the waveform will be distorted or clipped. The assumption in the tests up to now has been that the surplus transistors are in the small signal group.

R.F. TESTS

All the transistors available should be put through the r.f. test, irrespective of whether they have passed

the previous tests. The r.f. test circuit is shown in Fig. 5. Here the reader will have to arrange for the crystal, Tr1, L1 and C1, according to what he has available, to check frequency. As a guide, with a 20MHz overtone crystal, an OC170, L1 being eight turns of 18s.w.g. on a ¹2in former, tuned with a 50pF capacitor, the circuit will cover the range up to

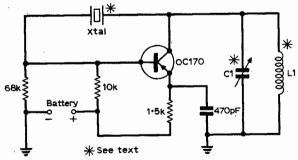


Fig. 5: The r.f. test circuit. For NPN devices the battery polarity must be reversed.

30MHz. Here a receiver or G.D.O. is necessary to detect the oscillation. Once oscillation is established it is merely a matter of substituting as before.

With your collection graded, as outlined, it is simply a matter of substitution of various types within that grade for optimum results in a particular project. The methods described are by no means comprehensive but at least you will have got a little order out of chaos and the building of many projects will be made much easier.

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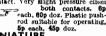
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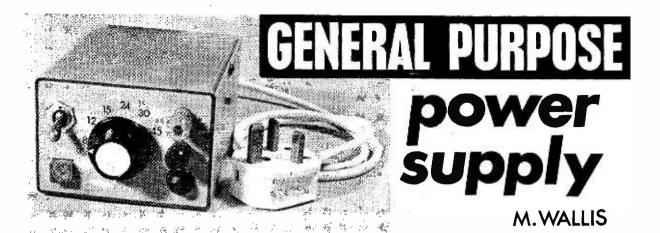
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'N the "good old days" of valves, power supplies were simple. Some 90 per cent of equipment operated with an h.t. supply of about 300V and the heaters were nearly all 6.3V. A single power supply would operate the vast majority of equipment that one was experimenting with. Today the semiconductor rules the day and power supplies are very much less standardised. Checking through a recent copy of Practical Wireless showed that there were six circuits given-and five different supply voltages were specified, ranging from 6V to 30V. Batteries can cope with a number of these, especially the most common 9V supply, but even the most half-hearted experimenter needs quite a range of batteries to pursue his hobby-even for quick lashups.

One can of course build a sophisticated, stabilised power supply with variable output covering quite a large range but such circuits are expensive, are by no means simple in their construction and most of the time the sophistication is not required.

By incorporating a simple two-pole, six-way rotary switch into the circuit of a standard power supply, it is possible to build a power unit with six different outputs—all with low internal impedance—and which will simulate exactly the supply that you finally build into the equipment.

A few refinements can be incorporated into such a power supply such as an indicator neon and fuse protected output which are not always necessary on the final supply which is fitted permanently but which make life simpler (and cheaper!) while experimenting.

the circuit

The wide proliferation of supply voltages needed for semiconductor designs has brought about the introduction of mains transformers with a wide range of secondary tappings and it is simply by selecting the appropriate ones of these that various output voltages can be obtained.

The transformer specified in the components list is a good one to use but other types with multiple secondaries can be adopted.

This one is fitted with 12, 15, 20, 24 and 30V tappings (these of course refer to a.c. r.m.s. volts)

VOLTS (d.c.)	VOLTS (a.c.)	Connections from which derived
4-5	3	12—15
6	4	20-24
7-5	. 5	15-20
9	5 6	24-30
12	8	12-20
13-5	9	15-24
15	10	20-30
18	12	0-12 or 12-24
22-5	15	0-15 or 15-30
27	18	12-30
30	20	0-20
37.5	24	0-24
45	30	0-30

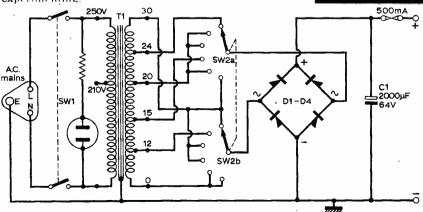


Fig. 1: The circuit of the general purpose power supply.

and a total of 13 different voltages can be selected from this as shown in the table. Those shown in the drawings, and for which the wiring is done, tend to be the higher voltages as batteries are still the best bet up to about 9V but it is a matter of personal choice which are selected.

Note that the d.c. voltages are those which should be read with no loading and are only approximate. In any case it has been found by

* components list

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experience that the secondary taps are by no means always accurate and quite large differences have been measured from the theoretical voltage but as exact voltages are rarely needed, variations are allowable.

The output voltage from the prototype, which can be seen marked on the chassis, are those measured, not those calculated, and these discrepances can only be due to winding inaccuracies as it is wired up exactly as in Fig. 1.

The current rating required from a general purpose power supply will depend on the uses to which it is put. Only power amplifiers at peaks take more than about 500mA and that was the rating chosen; it is sufficient for nearly all experimental work

More than one company makes transformers with the voltage taps used here, but the most widely available is possibly the Douglas MT112AT. The same transformer with a 2A rating is coded MT3AT and although it will just fit into the chassis shown, it is a tight fit and a larger one is recommended.

The mains supply is connected to the correct primary (either 210V or 250V) of the transformer via a double-pole switch. A panel neon, connected

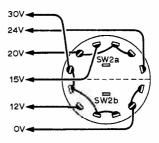


Fig. 3: The switch wiring used on the prototype.

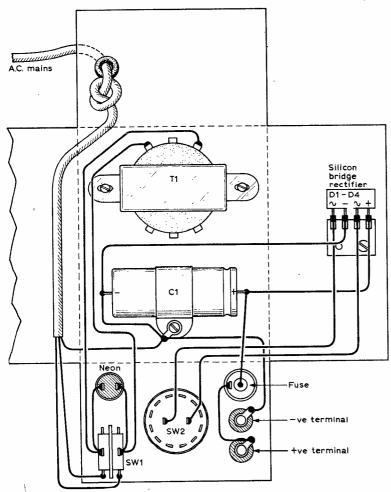


Fig. 2: The component layout and wiring details.

across the primary is very useful. Today most neons are supplied with a built-in resistor but if one is not fitted, a $220 \mathrm{k}\Omega$ should be wired in series as shown in the circuit.

The earth line is shown connected to the negative line but may equally well go the positive if used in conjunction with PNP type equipment which is earthed by some other means. Whatever happens the aluminium chassis should be earthed for safety reasons.

The various connections to the voltage selector switch are taken from the appropriate tappings and the switch wipers are connected directly to the bridge rectifier a.c. contacts. Plastic encapsulated silicon bridge rectifiers are widely available at reasonable prices and come under a variety of codings. The ratings must be 50V peak, 1A. The Mullard BY164 is suitable here.

Adequate smoothing for all purposes is provided by the $2000\mu F$ capacitor; the voltage rating of this should be well above the 45V or so applied and a 64V type was used in the prototype.

To provide a measure of protection, a 500mA fuse should be connected in the supply as shown.

construction

The prototype was built in a

small aluminium chassis size $5 \times 4 \times 2^{l_2}$ inches, fitted with a drop-in lid. These chassis are available from H. L. Smith Ltd., the address is given in the components list.

The layout of the components is shown in Fig. 2

and can be seen in the photograph.

The bridge rectifier can be mounted on a small tag strip on one long side, the transformer and the

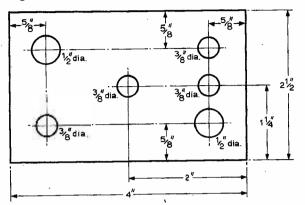
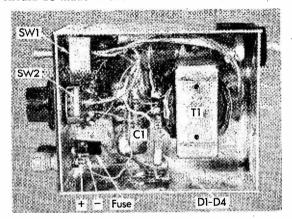


Fig. 4: The front panel drilling details.

capacitor on the base, with the remaining components on the front panel.

The switch wiring needs quite a lot of planning unless that used in the prototype is copied exactly, this is shown in Fig. 3. If other voltages, selected from the table, are required, the wiring should be planned carefully first of all. It is far easier to do the wiring on the switch leaving a few inches of wire which will finally be wired to the transformer.

When all wiring is completed a thorough check should be made to see that there is no possibility of



Internal view of the prototype showing component siting. Compare this with Fig. 2.

mains being applied to the case and that no bare wires can touch each other.

Connect a voltmeter to the terminal sockets and switch on, ensuring that a voltage reading is obtained in each switch position. Wrong wiring on a switch of this sort can lead to a direct short of the secondary and—unless noted very quickly—unpleasant things can happen to both the transformer and the switch.

There may be a very small amount of sparking while switching but this does no harm. The prototype has been in use continuously for several months with no ill effects.

There is just one point that must be made. The charge stored in a smoothing capacitor of this size lasts for a long while and if the power supply has been on a high range the capacitor should be shorted out (with the mains switch off) before being applied to a circuit where a high voltage could cause damage. This should be done through a resistor, not directly, as the fuse may blow otherwise.

If, due to the nature of the experimenting, a considerable amount of switching is envisaged, a resistor can be wired permanently between the output terminals so that a small current is drawn at all times and the capacitor will then settle down very quickly to the applied voltage. A 4·7kΩ resistor with a ½W rating will serve this purpose.

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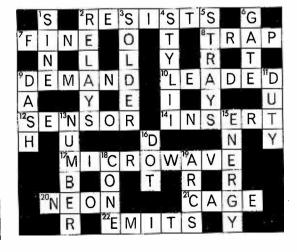
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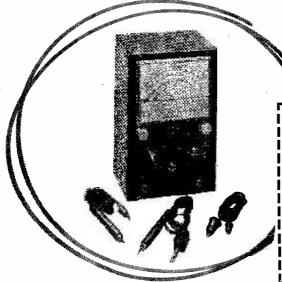
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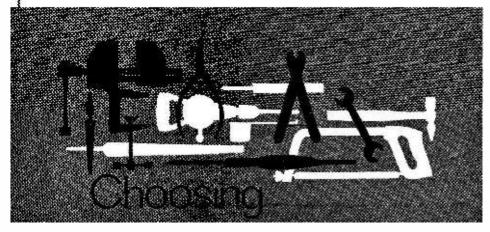
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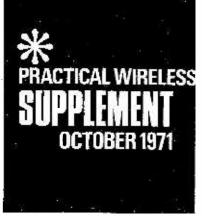
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SMALL WORKSHOP TOOLS S. YEATMAN

SINCE YOU ARE reading Practical Wireless the chances are that you have already tried your hand at constructing one of our projects. If you are already well fitted out with tools or have access to workshop facilities the bending up of a piece of aluminium to make a chassis or the punching and drilling of various holes would not have presented you with much of a problem. But if you are a newcomer to home electronics you have probably employed 'knife and fork' techniques, utilising kitchen knives and similar tools!

This really is doing things the hard way and the effort involved can put a considerable damper on anyone's enthusiasm for home construction, apart from any ensuing domestic problems when the lady of the kitchen discovers that her best kitchen knife now looks like a hacksaw blade!

If you are really taken with the idea of constructing electronic gear then be prepared to start off properly by buying a few good essential tools designed for the job in hand. Forget Dad's woodworking screwdriver with its $\frac{1}{2}''$ wide blade filed so thin that it seems to fit any screw head from 8BA upwards! Very convenient, of course, but what a mess it makes of screw heads! Dad's pliers, too, are very handy—they 'fit' any size of nut you like to mention but by the time they have slipped a couple of times that nut looks pretty sorry for itself.

When you find out what fun these electronic projects can be you will begin to take pride in the finished job and Mum's knives and Dad's tools just won't be good enough. So, start off on the right foot by getting the proper tools and remember that, like most 'bargains', cheap tools can turn out to be expensive in the long run. Spend a little more on tools from manufacturers of repute and scorn imported tools made of inferior materials.

The following review lists small workshop tools more or less in the order of necessity bearing in mind the requirements of the home constructor. The soldering iron takes precedence since even the simplest bought kit of parts needs the odd soldered joint or two.

In a review of this kind only general reference can be made to the great variety and range of tools that are available today. Specific recommendations are impossible without some knowledge of the amount of money it is intended to spend and the precise nature of the work to be undertaken in the workshop.

Workshops tend to 'just grow' from a few simple tools. As the limitations of these tools become apparent so more tools are bought to fill the gaps, thus it can be seen that a large initial outlay is not necessary.

Spreading the cost over a period allows emphasis to be made here on the requirement to buy quality, a policy which, if followed, will not be regretted.



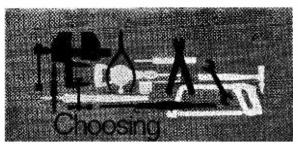
electric soldering irons available today, the main problem being the selection of one of a suitable wattage for the home constructor of radio gear. For the average Practical Wireless project a 25 watt iron should prove satisfactory but if a lot of work is likely to be done on printed circuit boards a 15 watt iron would be more appropriate. These circuit boards often contain many small components fitted quite close together and thus vulnerable to the heat of a soldering iron. Excessive heat can also damage the copper strips of the board itself.

With these irons goes a wide range of copper bits of differing diameters, some straight and some angled. Initially a straight bit $\frac{1}{8}''$ or $\frac{3}{16}''$ in diameter can be chosen for general purpose work with an angled bit for the awkward spots. At a little extra cost special bits are available which contain additives to the copper designed to prolong the life of the bit.

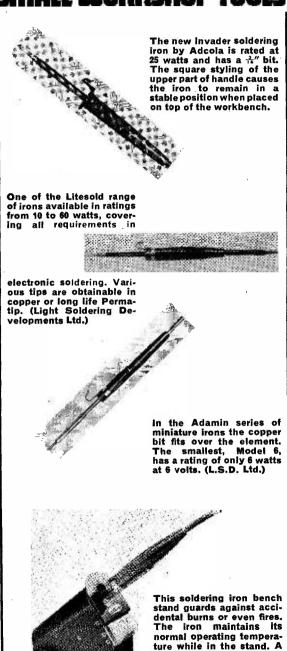
For home use a soldering iron operating directly from the 230 volt mains supply is usual but, for anyone who might feel a little happier with a lower working voltage, irons are available to work from as low as 6 volts via a mains isolating transformer.

When wiring an electric iron to its mains plug ensure that the instructions supplied with the iron are strictly adhered to so that the iron is properly earthed. This is a very important point for the builder of electronic equipment when wiring in transistors and other solid-state devices. Irreparable damage can be caused to these devices by stray earth currents if the soldering iron is not properly earthed.

Advancing rapidly in popularity are the soldering guns or 'heat guns', as they are sometimes known. The copper bit is, in fact, part of a low voltage secondary winding of the built-in transformer. Exaggerated claims are sometimes made concerning the short heat-up time for these guns,



SMALL WORKSHOP TOOLS



typically 10 seconds, but the better known makes can be relied upon in this respect.

The trigger style on-off switch on these guns often has two positions, for fast heat-up and for running heat, thus giving a dual-heat facility. Sometimes other types of bit are available which can be used to cut sheet plastics or to weld plastic sheet and so on.

Solder today, as used on electronic work, invariably has several cores of flux to ensure better distribution of the flux around the joint and consequently a better joint. The solder contains tin and lead in varying proportions which affect the temperature at which the solder will melt. The usual ratio of tin to lead is 60/40 for radio work and the diameter of the solder can be between 18 and 22 s.w.g. depending upon the fineness of the work being done.

While short lengths of solder can be bought for a few pence the ½ or 1lb. reel will be found to be more economical for the home constructor. If however he wishes to experiment with various solders he will find many easy-to-use dispensers and handy-packs available with different types of solder such as 40/60 or 60/40 tin-lead alloys as well as solders containing a small amount of copper which is calculated to reduce the amount of wear on the copper soldering bit.

Although soldering irons are usually fitted with a hook of some kind so that they can be hooked onto a convenient point it is a good policy to provide some kind of metal stand for use on top of the bench. Such a stand can be easily made out of an odd piece of aluminium or tinplate.

Soldering iron manufacturers naturally produce their own stands or 'protective shields' for their irons and these are designed to ensure that the iron does not overheat when the iron is not being used but is still switched on. For convenience a cleaning pad for the bit may be incorporated and possibly a holder for a reel of solder.

A 'must' in soldering is a sharp pick or long needle for clearing out holes in soldering tags and removing surplus solder. Commercial versions are usually made of aluminium, to prevent the solder sticking to the needle, and with a plastic handle or covering. One end may be pointed and the other end split to enable wires to be unravelled from tags etc., while the solder is molten.

Another de-soldering aid available is in the form of copper wire braid. When the end of the braid is touched on a hot soldered joint surplus solder travels up the braid by capillary action. The end of the braid is snipped off ready for the next joint.



PLIERS AND CUTTERS. Again a very large field of choice and again the rejoinder to pay a bit extra to get quality. Two pairs of pliers will suffice for most constructional work, a long-nosed pair with tapered jaws coming to a reasonable fine point and a heavier pair with stocky jaws, maybe $\frac{1}{2}$ wide, and probably incorporating a pair of sidecutters.

The long-nosed pliers will always be in the hand of the electronics constructor, for twisting wires round tags and forming the wire ends of components before soldering them into position. Don't forget that a wiring joint should always be mechanically sound before it is finally soldered, although the true experimenter will not go along with this principle because he will want to use the same components time and again!

Variations on the pliers theme include those with angled

bit cleaning pad is fixed to

(L.S.D. Ltd.)

front of the stand.

jaws of various lengths, round jaws and flat narrow jaws rather than pointed ones. When buying pliers see that the opening and closing action is not too loose or the jaw tips will not close together properly and it will be difficult to get hold of fine wires when wiring up.

Wirecutters will be used as often as the pliers so, again, only the best must be chosen. They should be about the same size as the long-nosed pliers with the cutting edges at an angle to the body. Other types have the cutting edges across the top or parallel with the body.

When buying sidecutters hold them up to the light and see that the cutting edges come together cleanly and evenly throughout their length. Try this test with a cheap imported pair of cutters and see the difference! The previous remarks concerning the action of pliers applies just as much to sidecutters. A loose action and the cutting action will be greatly impaired. On the other hand, there is nothing more annoying than a pair of cutters or pliers that need two hands to open them!

WIRESTRIPPERS. Different people have different ways of stripping the insulation from wire, varying from the use of the front teeth to an old kitchen knife! If the wire is nicked, and it often is, it will eventually break if it is subject to vibration or flexing. The ordinary sidecutters are frequently pressed into service as wirestrippers mainly because they will have just been used to cut off a piece of wire and are still in the hand! However there are plenty of wirestrippers on the market designed for the job so there is no excuse for improvisation.

Wirestrippers usually have some means of adjustment to cater for different size wires. It is important not to try to strip wires of a larger gauge than that for which the wirestripper has been set, or the wire will be nicked. One of the latest style wirestrippers on the market has a four position cam that can be adjusted immediately to another gauge of wire, while still held in the hand. Incidentally, this model also has wire cutters which have ground edges, a great improvement.

A feature on some types of wirestrippers, which the author considers of great importance, is the provision of a spring that keeps the cutters normally open, the cutting action taking place against the spring.

SCREWDRIVERS. Strictly speaking, every different sized screw needs its own screwdriver, if the blade is to fit the screw slot properly, but not many of us are likely to go to that extreme. The width of the blade should be about the diameter of the screw head or wider, rather than narrower. For electronic construction work it is best to buy a set of screwdrivers with blade widths from about $\frac{1}{6}$ " up to $\frac{1}{4}$ " or $\frac{5}{16}$ ". Three screwdrivers should be sufficient. A useful addition is one for the Philips style screws, that is the screws with crossed slots.

Screwdriver sets are available consisting of several blades that can be fitted to a common handle. Although these may be cheaper than individual screwdrivers they are less convenient in practice.

The screwdriver blade itself should be about 6'' or so long depending upon the blade width. One very useful size of blade is one about $\frac{3}{32}''$ wide for use with knob setscrews. The blade should be as long as possible so as to reach a knob in spite of the clutter of other knobs, dials etc. on a panel.

Don't take any notice of the markings on some screwdrivers that seem to indicate that one can use them with safety on 10000 volt circuits! Anyone having that sort of voltage around should know the precautions he should take. They would not include the use of a screwdriver on 10000 volts!

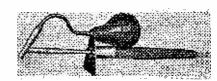
Variations on the basic screwdriver include the ratchet

A most useful de-soldering aid is a reel of specially treated copper braid which will completely remove unwanted solder from a joint by capillary action.

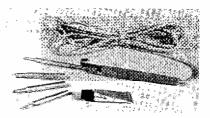




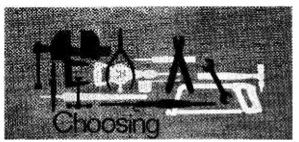
The Adcola protective shield is intended for wall or bench mounting and has slots for a spindle carrying a reel of solder.



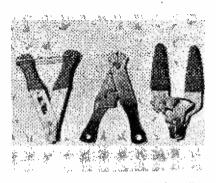
This de-soldering tool (L.S.D. Ltd.) could be invaluable when stripping components from surplus equipment. It is self-contained and simple to operate and may be used as a conventional soldering iron.



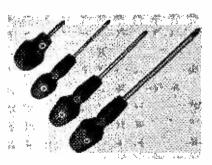
The Adamin miniature iron, Model 15, can be obtained complete with a set of four bits of differing size. The tube of special lubricant is used to prevent the bits from sticking to the heating element.



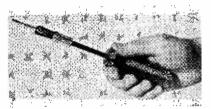
SMALL WORKSHOP TOOLS



Three of the Bib range of wirecutter, strippers by Multicore Solders Ltd. The latest, centre, is the Model 3A with an instantly adjustable cam enabling the strippers to deal with most Wire sizes. The cutters have ground cutting edges. The ends of the arms are, in fact, flat spanners for 4 and 6BA nuts, a most unusual feature.



A set of Stanley 'Pozidrive' screwdrivers enabling Philips type screws to be reached in the most awkward corners.



The Stanley range of 'Spee-D-Grip' screwdrivers incorporating a spring clip which holds screws in position for starting in inaccessible places.

types which remove the necessity of continually lifting the screwdriver from a screwhead thus, incidentally, reducing the risk of the blade slipping and doing damage. Of course, if one used a screwdriver properly, the blade should never leave the slot, but this seldom happens!

A most valuable addition to a set of screwdrivers is one having a spring attachment at the tip of the blade to firmly hold a screw while it is being inserted into place. Its value in awkward spots is inestimable. Such an attachment can sometimes be bought separately for use on almost any screwdriver.

SPANNERS. There is always a temptation to use a pair of pliers on any nut that needs tightening, usually because the pliers are ready to hand. It should be remembered that the opposite faces of a nut are parallel to each other where-as the jaws of a pair of pliers are only parallel when they are closed. That's why the pliers always slip off the nut and cause damage.

First of all, get a set of box spanners covering 0 to 8BA either single or double ended. The single ended ones with a wooden or plastic handle are recommended. Just two, for 4 and 6BA, are worth their weight in gold in a workshop. Having long stems they are invaluable for placing nuts into odd corners where fingers or pliers cannot reach.

It is always good practice, wherever possible, to tighten a nut on to a screw rather than use a screwdriver on the screwhead. This is particularly important where the screwhead is visible such as on the front panel of a piece of equipment where a damaged slot on a screw would be an eyesore.

Another requirement in the spanner line is a set of flat spanners, again covering the BA sizes, possibly 0 to 8BA. These can very well be double ended ones since the length of the spanners is not of much importance.

Some sets are single ended but bolted together at one end so that one can be selected and the others swung out of the way, rather like a set of feeler gauges.



DRILLS. In the past a hand drill would have been a first acquisition in a workshop, followed by an electric drill when the money was available. Today the cost of electric drills is such as to justify buying one from the outset.

The simple electric drill will have a single speed and accommodate drills up to ½" in diameter, ('twist drill' is really the proper name for the bit that does the cutting). This model will perform just about all the routine drilling jobs in a workship where electronic projects predominate. A chuck adaptor to take drills up to ½" in diameter is highly desirable, enabling the usual holes for mounting potentiometers etc. to be drilled without having to resort to files or reamers.

If the drill is likely to be used for other duties around the house then the two speed drill with a chuck taking drills up to $\frac{1}{2}$ " in diameter can be considered. A popular accessory for the electric drill is the speed controller which is an electronic device enabling the speed to be varied continuously from a few r.p.m. to maximum speed. Such controllers are available in kit form or ready assembled.

If a lot of drilling is anticipated, particularly on chassis work, a drill stand will prove a boon. Most electric drill manufacturers have such stands but their design is such that they will accommodate most makes of drill. Such

stands ensure that holes are drilled vertically, which may not be very important on sheet metal work but could be critical with a deep hole. Holes will be more accurate both in size and positioning when drilled on a stand rather than with a hand held electric drill.

When using drill adaptors enabling, say, a \S'' drill to be used in a chuck of nominal $\frac{1}{4}''$ capacity, remember not to overload the motor by exerting excessive pressure on the drill especially in hard materials.

Hand operated drills suitable for the small workshop range from the simplest single speed ones with a ½" capacity to those taking drills up to ½" and having an integral gearbox providing either a low speed or a high speed at the chuck. A little extra spent on a good hand drill will be a worthwhile investment. In particular the gears will run more smoothly and require less energy and the chuck will be of a better quality.

Now, what about the twist drills themselves? These are bought singly, as required, or, if starting more or less from scratch, then in a set complete with a stand for use on top of the bench. They may be had in sizes from about $\frac{1}{32}$ " to $\frac{1}{4}$ " in diameter or in a numbered range from No. 80 (0·0135") to No. 1 (0·228"). Letter drills carry the range on upwards from 'A' (0·234") to 'Z' (0·413").

For many years the author has had just four drills always to hand, No's 12, 24, 32 and 43. These will cover the drilling of holes for tapping or clearing for 2, 4 and 6BA screws.

As far as the quality of twist drills is concerned, it is necessary to take a little more care over their selection than with other tools since there are a lot of drills on the market that seem to be made of little more than soft iron! So once again beware of cheap imitations and if the pocket will stand it, buy a small set of genuine high speed drills. These will last for ever and seldom need sharpening if not abused.

PUNCHES. Whenever it is required to drill a hole in a chassis, or anywhere else for that matter, it is essential to use a centre punch to mark the spot where the hole is required. Otherwise it is almost certain that the drill will wander off the spot, scratching or damaging the surface being drilled. This particularly applies to hard surfaces and when using a hand held electric drill especially if care is not taken to keep the drill vertical.

There's not much to say about centre punches. A good one will be of tempered steel with a sharp round point, usually machine ground. Spring loaded punches are available which obviate the need for a hammer, so these punches are very convenient and the extra cost well justified.

Chassis punches are a necessity although the number bought for a small workshop will probably depend upon the funds available. These punches will cut holes ranging from \$\frac{2}{3}" to 3" in diameter but it is recommended that at least two be obtained, \$\frac{2}{3}" diameter for B7G valveholders and \$\frac{2}{3}" diameter for B9A valveholders. These punches require a suitable Allen key to operate them and their cutting limit is about 16 s.w.g. mild steel. They will cut through the usual 16 s.w.g. aluminium sheet like the proverbial knife through butter, thus saving a great deal of time and energy.

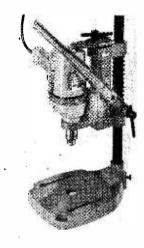
A useful tip...the resulting hole will have a burr on one side so place the cutting part of the punch on the side of the panel or chassis that might be of some importance from the point of view of appearance.

Other chassis cutters are available that will cut square or rectangular shaped holes but the infrequent need for such cutters does not justify their purchase.

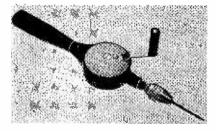
The predecessor of the chassis cutter was the tank cutter. This had an adjustable cutting bar for different diameters of hole, but unless used in a lathe or drill press,



This bench clamp enables a portable electric drill to be used with grinding, polishing or other attachments (Wolf). The drill is a Wolf Sapphire 73.



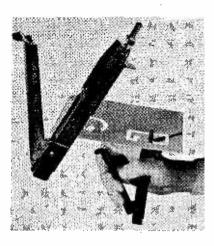
For precision drilling a bench drill stand is essential. The spring loaded press magnifies the hand pressure some ten times. (Wolf)



The old-fashioned hand drill now has a streamlined appearance. In this Stanley drill the gears are completely enclosed protecting them from dirt, swarf etc. and obviating nipped fingers!



SMALL WORKSHOP TOOLS



Requiring only a $\frac{1}{16}$ " hole to begin operations this Adel nibbling tool will cut a hole of any shape or size in sheet steel up to 18 gauge steel or aluminium up to 16 gauge. (West Hyde Developments Ltd.) The punch and die action cuts without distorting the metal.



This precision made Mole 'Supercut' tool has a powerful action capable of cutting long straight lengths, curves or angles in sheet materials such as steel, aluminium, plastic laminates etc. A replacement blade can be fitted very easily.



The standard Mole wrench has many uses in the small workshop. In spite of its immense gripping power it can be opened at a touch on the release lever. the resulting hole could be a little rough in appearance. A slow cutting speed was essential and for this a carpenter's brace was ideal,

Other hand tools for cutting holes in sheet metal include the family of "nibblers". A small hole is first drilled in the sheet metal, the nibbler inserted and the action of the tool nibbles the material away to form a hole of almost any shape desired. The hole can be finished off with the usual files.

Shears, or tinsnips, will be needed for cutting sheet metal and a 10" pair with straight jaws should suffice for general work. The will make a straight cut in sheet metal for a limited distance although the 'goosebill' type will usually prove better for this work.

TAPS. Now and again the necessity will arise to tap a thread in a hole and for this two taps for each size thread are required, together with a tap wrench. The first tap is tapered, to start the thread, and the second tap is full size. Sometimes a third tap of intermediate taper is used but this is not really required for small threads.

The tap sizes most likely to be required are the usual 2, 4 and 6BA. Taps can be broken off relatively easily and for this reason a proper tap wrench is strongly recommended.

The simplest and probably the best tap wrench is the bar type with a screw grip on to the tap. The chuck model is more expensive but it does enable the tap to reach positions that would be inaccessible with the bar type wrench.



FILES. For finishing off the edges of metalwork and suchlike a 10" flat file will be quite adequate. That known as 'second cut' will prove best being neither too coarse nor too fine for work on soft metals. A similar sized half round file is very useful for finishing off large holes such as those required for meters. While smaller holes, for valveholders etc., can be cut with chassis punches the cost of similar punches for large meter holes tends to be prohibitive, especially as they are used relatively infrequently. Such large holes can be made by drilling a circle of small holes, then knocking out the disc of metal and finishing the work with a half round file.

A 4" round file completes the essential complement of files. This will find most use in enlarging small holes in the absence of a reamer or for 'drifting' holes that are not quite where they ought to be! Perhaps the spot was not properly spotted with a centre punch before drilling!

With the advent into common use of slide switches on electronic equipment the need has arisen for a couple of small needle files, one round and the other square, for finishing off the rectangular holes required in panels for these switches. If a set of needle files can be managed then so much the better as these sets contain files of circular, triangular and flat sections as well as those already recommended.

When buying files get suitable wooden handles for them at the same time. Don't try to use a file without a handle or there will be a grave risk of the tang of the file entering the wrist if the file should jam or stick in the work.

HAMMERS. As far as hammers are concerned the small workshop should have one with a plastic or rubber head for bending sheet metal, without causing undue marking of the metal. A second hammer should be a light

one of the 'ball-pein' variety for general work around the place. A good make is advisable to ensure that the hammerhead is properly attached to the shaft as a hammerhead in orbit around the workshop is not funny!

REAMERS. During the construction of a piece of electronic equipment one always seems to need to drill a hole that is between the sizes of the drills available. In such a case a tapered reamer can be invaluable, as an alternative to filing out the hole with a round file, a process that does not endear one to one's family, if they are within earshot!

A pair of reamers can deal with holes between $\frac{1}{8}''$ and 1'' in diameter producing a clean circular hole with little or no burr. They are usually intended for use in a hand brace but can on occasion be held in the hand.

VICES. If a bench or table is permanently available for constructional work then a vice of some sort will prove to be a great boon especially when fabricating metal work. While smaller sizes are obtainable, an engineer's vice with 4" wide jaws should be the smallest size to be considered. Such a vice is extremely rugged and will take a lot of punishment, including using it as an anvil!

Since the vice jaws are made of hardened steel, and serrated at that, it is essential to make up a pair of soft jaws to protect work held in the vice. These jaws can be made from pieces of aluminium or other soft metal. Other vices have clip-on jaws which are plain, serrated or slotted to hold tubing or round material more securely. Another model of vice has a separate mounting plate that can be attached to almost any flat surface by means of a suction pad.

A carpenter's vice with 6" or 8" jaws is even better for sheet metal work since the jaws can be opened to a far greater extent than those of an engineer's vice. However the space needed to permanently mount a carpenter's vice, especially below and behind the vice, is considerably more than that needed for an engineer's vice. However, there are some portable vices on the market that seem to combine the best points of both types.

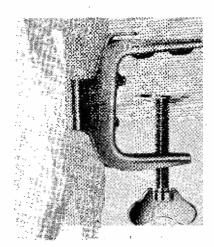
Although one may possess a bench vice there is always a need for a portable clamping or holding device and this is where the wrench comes in very handy.

The modern version of the wrench has an easily operated lever that allows the wrench to be locked on to the work and just as easily released. A clamp to hold the wrench is also available, which can be fitted to a table or bench top with a single clamping screw, thus providing a fixed vice if necessary. The jaw opening of the wrench can be adjusted over a wide range by a single screw fitting.

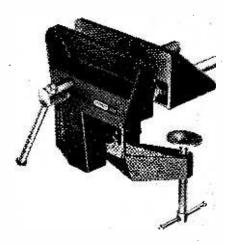
RULES. For marking out on sheet metal a rule and a square (trysquare) will be required, if an accurate job is to be done. A 12" rule is good enough and although it would seem that choosing such a simple tool would be no problem, the choice is, in fact, very wide indeed, ranging from rather inaccurate wooden rules, with printed graduations, to engraved steel ones.

Although a 6" square can be bought separately it really pays to buy a combination square since this incorporates a rule and square, usually of reasonable quality.

HACKSAWS. The hacksaw will come in for quite a bit of use in the small workshop, mainly for cutting sheet metal to size before forming into a chassis. Some of the cheaper hacksaw frames will be found to twist when the blade is tightened up, so a good quality frame will be a sound buy. The adjusting nuts, too, can be difficult to turn owing to poorly formed threads.



The versatility of the Mole wrench can be greatly increased by the use of this table or bench clamp, converting the wrench into a portable vice.



This Stanley aluminium portable vice has 5" jaws and can be clamped to any convenient surface. It is eminently suitable for bending sheet metal into chassis etc.



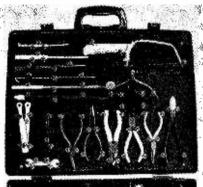
The Mini-Mole is just 5" long and very useful for getting into those tight corners, inaccessible to its 'big brother'. The curved jaws provide maximum contact with round or irregular shapes.

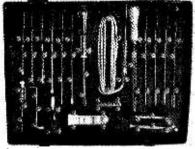


SMALL WORKSHOP TOOLS



This set of Philips trimming tools and adaptors will cover every conceivable requirement when aligning electronic equipment. The compact packaging makes the set particularly useful to the service engineer. (Combined Electronic Services Ltd.)





If you don't feel like tramping around the tool shops to gather together your collection of small workshop tools then this very comprehensive tool kit may be the answer. It has a wide variety of spanners, screwdrivers, pliers, cutters, a soldering iron and solder, and even an angled mirror for viewing components tucked away in tight corners. (C.E.S.Ltd.)

A frame to take 10" or 12" blades is big enough and the blades themselves can be of the low tungsten flexible type. These will deal with general cutting of mild steel, brass and aluminium etc. and absorb a lot of misuse without breaking. For maximum performance high speed flexible blades are to be preferred although they are a little more expensive.

For maximum efficiency the size of the teeth on the blade must be related to the material being cut. For soft metals, such as aluminium, a coarse blade is desirable to prevent the teeth from clogging, but on harder metals such as brass or on thin sheet metal a finer toothed blade is required. The number of teeth per inch of blade normally runs from 14 to 32 for hand held saws.

A very useful addition to the standard hacksaw is the 'junior' or bow-frame saw in which the blade is held under tension by the bow. The standard blade is 6" long with fine teeth. This saw is very handy for the smaller, finer jobs where the standard hacksaw would prove clumsy. In the electronic workshop the bow saw will be used a great deal for shortening the spindles of potentiometers, capacitors etc. Remember to hold the unwanted end of the spindle in the vice while cutting, rather than the component itself.

TRIMMING TOOLS. Although perhaps it is wandering into the realms of servicing it is not inappropriate in this review to mention trimming tools for electronic equipment. After all, having built a lovely radio set in our nice new workshop, it won't be of much use unless the tools to align it are to hand.

These trimming tools can be bought individually, as needed, but preferably they ought to be obtained as a klt. This will prevent the use of ordinary screwdrivers for adjusting iron cores which can only cause damage in the long run.

A kit may consist of one basic tool holder with several types of trimming tool that can be fitted into the holder or, in the more expensive kits, a complete trimmer for each purpose, such as adjusting iron cores, capacitors and special types of capacitors, such as the Philips pre-set beenive capacitors, which are met with quite frequently.

In conclusion it should be pointed out that for every basic tool mentioned in this review there are a dozen or more variations on the market. Like women, there are short ones, tall ones, thin ones and fat ones! Each one to his own fancy! The average D.I.Y. or tool shop is sated with gadgets that promise to relieve us of the tedium and work required to do a particular job. Few succeed, and those that do probably take twice as long to set up to do the job as one would take using basic tools alone.

It's surprising how often one resorts to basic methods and tools in the workshop. If funds are limited then put them to good use by buying only the best in tools from reliable manufacturers. If their names happen to be a household words they would not have reached those dizzy heights by selling rubbish.



SMALL WORKSHOP TOOLS

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82/450V	20p	8+16/450V	20p	82+32/450V	88
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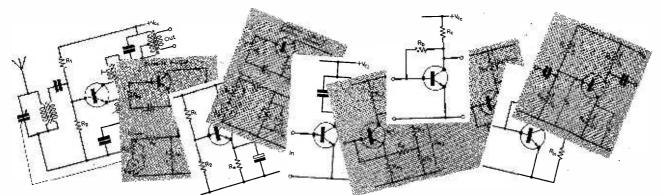
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TRANSISTOR CIRCUITRY for beginners'

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HEN the idea for this series was first broached, everything seemed straightforward and simple. We would entitle the wallchart 'Building Bricks,' said the Editor, in an inspired moment, and the articles would take up each 'brick' and develop it so that newcomers to transistor circuitry could understand what was happening inside those inscrutable black boxes.

Which was all very fine—but it completely overlooked the fact that any beginner worth his salt is

going to say 'But Why???'

No good my declaring '... an input signal to the base of Tr1 turns it on, which causes Tr2 to turn off and Tr3 and Tr4 to return to the stable mode, releasing Tr5, which now bottoms and allows ... etc.' It would sound like a paragraph from Henry!

To understand transistor action, it is imperative that we begin by understanding what makes the darned things tick, That means going back to pure physics, not too deeply, preferably without mathematics, and then progressing to the methods of construction and the reasons why there are essential parameters—rules of operation, if you like—and this is what we shall begin with.

Semiconduction

As the name implies, the semiconductor is neither pure insulator nor electrical conductor, but something in between. Its resistivity lies roughly midway between that of conductors and insulators. This resistivity, moreover, decreases with increase of temperature (the resistance of a conductor increases slightly as the temperature is raised). This is one important difference between semiconductors and conductors.

Next, perhaps more important, is the way in which current flows in a semiconductor. There can be two processes; either a flow of negative electrons, as in the case of a conductor, or a flow in the opposite direction of positive 'holes'. The concept of hole flow is so vital to the understanding of transistor operation that we must begin with a simple discussion of atomic structure in order to get a grasp of it.

Atomic Structure

An atom consists of a central nucleus containing positively charged protons with a number of negatively charged electrons orbiting around it. The negative charge carried by an electron is equal to, and exactly opposite to, the positive charge of a proton. In a neutral atom, the numbers of each are equal.

Different elements have atoms containing different numbers of electrons, orbiting at different distances from the nucleus. The electrons close to the nucleus are tightly bound to the nucleus, but those circling in the outer orbits are more loosely bound, and are easier to dislodge from the parent atom. The outer ring is called the **valence ring**, and the valence electrons are those with which we are mainly concerned in this work. The looser they are bound to the atom—in any material—the better a conductor a piece of that material will make. This is because loosely bound valence electrons are free to be attracted by a positive charge, such as a voltage applied across a conductor by means of a battery.

Electrons flow out of the conductor into the battery at the positive connection, but an equal number pass from the negative connection of the battery into the conductor at the negative terminal, keeping the overall total of electrons the same.

Conversely, an insulator is a material in which all electrons are tightly bound to the nucleus, so that a large amount of energy is needed to break any loose to make a current flow.

Crystal Structure

The structure of a crystal of germanium or silicon is depicted in Fig. 1. It can be seen that there are four valence electrons in the outer ring and these are balanced in the nucleus by four positively charged protons, as indicated. In the three-dimensional structure that exists in reality, every atom is equi-distant from four other atoms and each valence electron forms a pair with one from an adjacent atom.

In two-dimensional form this can be shown as in Fig. 1, with pairs of electrons forming co-valent

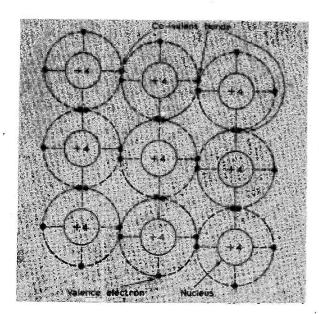


Fig. 1: Illustrating the structure of a germanium or silicon crystal.

bonds. This is a perfect crystal, with no impurities; and no electrons are dislodged from their atoms. This is, in fact, the basic structure of a perfect insulator, at absolute zero temperature, but as temperature increases, the energy of each atom is increased and some electrons can break away. If a voltage is applied, these free electrons are attracted to the positive pole and a current flow results. The space left in the crystal structure by a dislodged electron is termed a 'hole' and this represents a net positive charge.

We can therefore think of a movement of negative electrons toward the positive voltage terminal and a virtual movement of positive 'holes' (as successive electrons move to fill the vacant spaces) toward the negative pole.

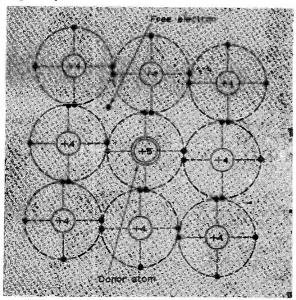


Fig. 2: Adding a 'donor' atom producing an excess free electron.

N-type crystal.

N- and P-Type Semiconductor

This is a simplified explanation, but suffices to show the principle of hole movement. In a pure semiconductor there must be one hole for each free electron. Such a pure semiconductor is termed an intrinsic material. However, if an element of impurity is added to the intrinsic material an imbalance is obtained. If, for example, a "donor atom" having five valence electrons is added to the crystal, as in Fig. 2, there is an excess free electron. Four of the donor atom's valence electrons form co-valent bonds, in the manner previously described, but the free atom cannot do so and orbits loosely.

Some materials with a valency of five are antimony, arsenic and phosphorus. These can be added to the germanium or silicon crystal in proportions of as little as one part in a hundred million and reduce the resistivity of the germanium from 70 ohms/cm to 1 ohm/cm. Because the donor atoms in such a material contribute free electrons, which carry a negative electric charge, a material "doped" in this

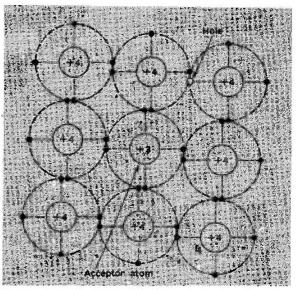


Fig. 3: In this case, an 'acceptor' atom creates a 'hole'. P-type crystal,

way is called an **n-type** semiconductor (i.e. it possesses an excess of negative current carriers).

The opposite process takes place if indium, borium or aluminium atoms are added to the basic semiconductor material, germanium or silicon. These materials have three valence electrons, and the addition of them creates holes in the semiconductor material, as shown in Fig. 3. The resultant semiconductor material has an excess of positively charged holes and is termed **p-type**.

Minority Carriers

Minority carriers are another important factor of semiconductor physics. If a few electrons are injected into p-type material they are called minority carriers (the majority of current carriers in p-type material being, as we have seen, holes). Recombination takes place as the electrons move to fill the holes.

Similarly, holes injected into an n-type material

form minority carriers in this type of material, and small currents arise as recombination takes place. The time this takes to happen, termed the minority carrier lifetime, is an important property of a semi-conductor material and the operation of most transistors depends on the movement of minority carriers.

PN Junction

A single piece of p-type or n-type semiconductor material is purely resistive: reversing the applied voltage does not alter the current flow. But if p-type and n-type regions are formed in a piece of semiconductor material, with the regions adjoining each other, a rectifying device is formed: when voltage is applied in one direction, current flows, but in the other direction little or no current flows. Such an arrangement is termed a pn junction.

In Fig. 4, the region to the right of the centre line is n-type, possessing a certain density of free electrons. To the left of the line is a region of p-type material possessing a number of positively charged holes. When the junction is first formed, it becomes an area of activity for a short period as free electrons tend to cross into the p-type region and holes move over to the n-type region.

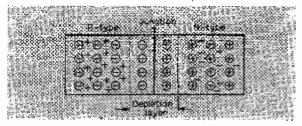


Fig. 4: Showing the charges existing in an unbiased PN junction.

This process of diffusion results in a potential building up across the junction, as the p-region gains a net negative charge with respect to the n region in the vicinity of the junction, and vice-versa, and this potential prevents further diffusion. This potential is termed the barrier potential and the region in the immediate vicinity of the junction, where there are very few free current carriers remaining after the initial diffusion, is termed the depletion layer or region, as shown in Fig. 4.

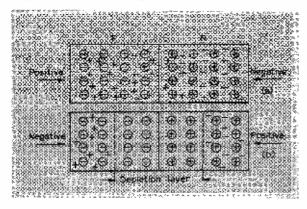


Fig. 5(a): Position of the charges with forward biasing and Fig. 5(b): with reverse biasing.

Forward and Reverse Biasing

We now have a piece of material with pronounced non-linear characteristics: when an external voltage is applied, current will flow in one direction only. If the p-type side is made positive with respect to the n-type (see Fig. 5(a), forward biasing), the effect is that the barrier potential is reduced. Majority carrier holes will diffuse across to the n-type (where they become minority carriers), and majority carrier electrons will move in to the p-type region. A relatively large current can flow across the junction in this way.

Reversing the bias, i.e. applying it as shown in Fig. 5(b), increases the barrier potential (since the external voltage adds to that across the depletion layer already) so that current flow is made very low—in fact virtually nil. However, due to the effect of temperature a number of holes and free electrons will be generated on either side of the junction—this happens even at room temperatures—and this will give the effect of a small reverse current flowing across the junction. The forward current may be quoted in milliamps but the reverse current only in microamps.

Characteristics of PN Junction

Fig. 6 shows the characteristics of a pn junction. As the forward voltage is increased from zero, the forward current increases slowly at first until the junction barrier is overcome, then rises rapidly. (In alternative terms, it can be said that the resistance of the forward biased junction is low.) In the reverse direction, the current remains low and practically

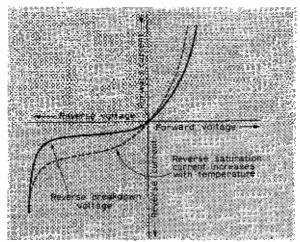
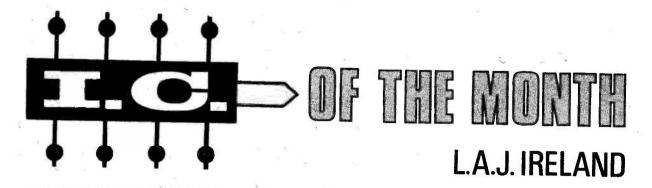


Fig. 6: The electrical characteristics of a PN junction.

constant as voltage is increased, i.e. reverse resistance is very high. Eventually, however, a point is reached where the reverse voltage accelerates the reverse current carriers to such a speed that they break down covalent bonds in the crystal structure, thereby releasing more carriers and producing an avalanche breakdown (when the junction may be destroyed). The value of voltage at which this occurs depends on the resistivity of the material.

As the reverse saturation current is due to the effect of heat, it increases with temperature, giving

-continued on page 512



Number 24

LM373 Multiple-mode Detector

Motorola MC1596 balanced mixer integrated circuit and outlined its operation as a multiple-mode detector in receiver circuits. The unit discussed now goes further; full i.f. amplification facilities, at any chosen frequency up to 12MHz., are incorporated, as well as multiple-mode detection. It is the National Semiconductors (U.S.A.) type LM373, available in the U.K. from: Athena Semiconductor Marketing Co. Ltd., 140 High Street, Egham, Surrey: or Rastra Electronics Ltd., 275 King Street, Hammersmith, London W.6.

Uses

The unit is intended, not just to operate in each mode, but to make it possible to design a receiver in which intermode switching would be simple without reduction of performance. As well as the standard a.m. and f.m. modes, as used for broadcast entertainment in the m.f. and v.h.f. bands, the LM373

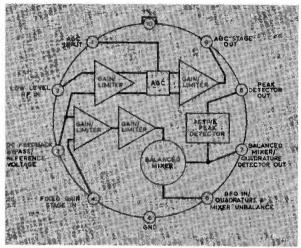
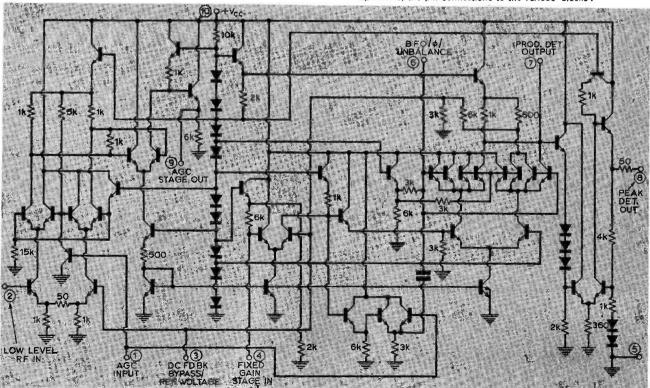


Fig. 1: The equivalent circuit of the LM373 (below) with (above) the top view of the pin connections to the various 'blocks'.



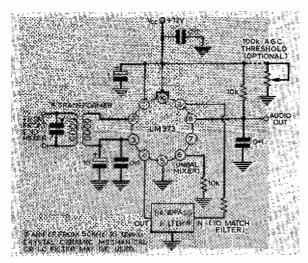


Fig. 2: External circuitry required for the LM373 when used as an AM IF strip.

can handle n.b.f.m. (narrow-band frequency modulation) increasingly used in mobile radiotelephone systems, and s.s.b. (single sideband suppressed carrier) transmissions of commercial or amateur origin.

Design

Considerable ingenuity is exercised in the design of the i.c. to reconcile the conflicting demands of each transmission system with the minimum of intermode switching in the receiver. A.M. and s.s.b. receivers must have a linear amplitude response, maintained, despite the wide range of signal strength received off the air, by an automatic gain control subcircuit. An f.m. i.f. amplifier strip, on the other hand, is essentially nonlinear, in that any input signal is amplified to a limiting level before application to the detector stage. It follows that automatic gain control has no place in such a system.

To provide the facilities indicated, the LM373 incorporates four gain blocks, which can be operated in a limiting mode, as well as an automatic gain

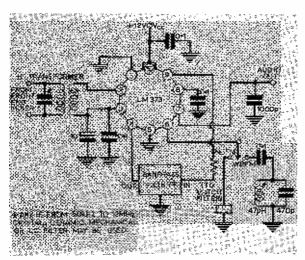


Fig. 3: The LM373 in its role of an FM IF strip suitable for narrow or wideband reception.

control system and two detectors, one a balanced mixer similar to that in the Motorola circuit already mentioned, and the other a peak detector for a.m. operation. The equivalent circuit and pin diagram given in Fig. 1 allows these sections to be identified.

Operation

Some discussion of the circuit would appear necessary. Each gain block is essentially the familiar differential amplifier, which can be operated in its linear region when the circuit is used for a.m. or s.s.b. reception, or driven to full gain as a limiter for f.m.

The automatic gain control characteristic is secured through a pair of transistors shunting the differential interstage coupling components. When no a.g.c. voltage is applied, these transistors are cut off, but as the control voltage is increased, they draw off an increasing fraction of the signal current; finally the point is reached where the signal from

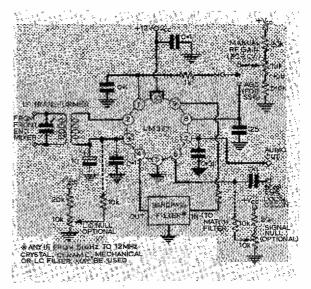


Fig. 4: The SSB/CW IF strip shown here further demonstrates the versatility of the LM373.

the first gain block is shorted out, and coupling to the second is merely through stray capacitance.

This approach secures the unusual a.g.c. range of over 70dB., enabling linear a.m. detection to be attained with input signals from 40 microvolts up to 200 millivolts. The string of reference diodes and emitter follower transistors from the d.c. +ve. line provide power supplies with decoupling to each stage of the amplifier strip, together with appropriate bias voltages to set the operating point of each element. As well as simplifying the external decoupling requirements of the circuit, these imply that each stage essentially has its own regulated supply, so minimising the effects of power line variations.

The balanced mixer stage is easily identified by the cross-coupling of its component transistors. For f.m. it operates as a quadrature detector, a system outlined when describing the MC1496, as was the product detector system used for s.s.b. In a.m. operation, this section is deliberately unbalanced by an externally applied bias, leaving only one side of

the differential system connected to its load effectively.

N.b.f.m. demodulation requires a quadrature element of a higher Q factor, i.e. more rapid change of phase with frequency, and therefore a quartz crystal is substituted for the more familiar l.c. circuit. The final section is the peak detector, acting as a.m. detector and a.g.c. voltage source for linear a.m. and s.s.b. operation.

A few precautions are necessary for successful operation. The a.g.c. stage output at pin 9 has a very low impedance, being taken from an emitter follower. A value of 70 ohms is quoted, so that if a high Q crystal filter is employed here a series resistor is advocated, so that the resonance peak of the component will not be damped. The manufacturer of the crystal in question will quote a recommended source impedance, allowing selection of an appropiate resistor.

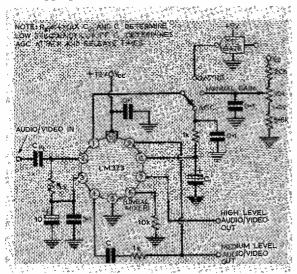


Fig. 5: In TV circuitry the LM373 can be used as an audio/video amplifier with a.g.c., manual gain control or gating.

The d.c. feedback loop at pin 3 must be carefully bypassed, since the decoupling must be effective throughout the passband of the LM373; an electrolytic or tantalum capacitor, as used for a.f. decoupling, has too high a reactance at high frequencies to be effective alone, so a small disc ceramic or silver mica is added.

For s.s.b., bypassing of r.f. at pin 7 is particularly important, as any local oscillator component reaching the peak detector would increase the a.g.c. voltage generated, and lead to an unnecessary reduction of system gain. The optional nulling circuits may be used to maximise gain should that prove necessary (an unlikely event).

Figs. 2-5 indicate the circuitry in which the LM373 may be employed, and it should be noted that the communications receiver applications for which it was designed do not exhaust the potential of this versatile unit, as is indicated by the circuit for a wideband amplifier indicated in Fig.5. Other reported applications have included use as an amplitude-modulated r.f. oscillator, an s.s.b. generator, or indeed any of the applications open to the balanced mixer circuit in which extra wideband gain is useful. So presumably more will be heard of the LM373!

TRANSISTOR CIRCUITRY FOR BEGINNERS

-continued from page 509

the effect shown by the dotted line in Fig. 6.

Junction devices of this type are known as diodes, and widely used in radio, television and other electronic applications. The diode is essentially a one-way valve, as we have seen from the forward and reverse bias characteristics.

Germanium diodes were the first types developed for commercial use, but the current-handling capacity of these devices was limited, and it was not until research provided a means of treating silicon that power rectifier semiconductors became a practical proposition. Silicon is more difficult to melt and purify than germanium, but can be doped with donor and acceptor impurities in the same way as germanium, is less sensitive to temperature, and will handle higher powers more efficiently. Common types pass currents of half-an-ampere and more with a practically negligible forward resistance. Silicon diodes have a high reverse voltage rating. (On the other hand, a germanium has greater efficiency at low voltages and at higher frequencies, and is thus retained for signal circuits.)

The pn junction forms the basis of most forms of semiconductor diode and transistor. One other form of construction is important in connection with diodes—the point contact type. In this the rectifying property is possessed by a contact between a thin metal wire and a piece of semi-conductor material. Though junction devices predominate today, point contact diodes are still much used as detectors in radio equipment. Their characteristic is similar to that of the pn junction, shown in Fig. 6, except that as the reverse current increases beyond the reverse breakdown point, reverse voltage tends to fall.

TO BE CONTINUED

TELEVISION

SEPTEMBER

TV INTEGRATED CIRCUITS

Integrated circuits are beginning to appear in the new TV chassis in increasing numbers, particularly in colour receivers. The types used have been specifically developed to fulfil the signal processing requirements of TV sets but nevertheless bring with them numerous new techniques. In a new series starting this month we shall be examining these new i.c.s. to familiarise readers with the various changes in TV set design they involve.

SERVICING TV RECEIVERS

Recent articles have concentrated on fairly upto-date chassis. This month however we are going back to a widely distributed "625-line convertible" chassis, the Sobell ST195/ST282 series. This will give us an opportunity to examine the faults common in older sets.

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SERVICEUS SERVICES AN INTRODUCTION TO FAULT-FINDING

PART 6 G. J. KING

THIS MONTH G. J. KING RETURNS TO THE SERIES TO ELABORATE ON AUDIO MATTERS, DEALING WITH CONVENTIONAL VALVE AMPLIFIERS AS WELL AS TRANSISTORISED HI-FI EQUIPMENT.

OW that colleague Hellyer has cleared the decks on alignment, superhet theory and f.m. stereo, it is my turn to have a go on audio equipment.

GRAM AUDIO STAGES

An ordinary radiogram differs little at the preaudio stages from a radio receiver, so much of the information presented earlier will also be applicable here. However, most contemporary radiograms are equipped for f.m reception, possibly with a decoder to abstract the stereo information and to direct the left and right channel signals separately to the corresponding audio channels. Thus, many radiograms and record reproducers feature a pair of audio channels and speaker systems. These are very much twin partners, the design of one following very closely the design of the other.

A mono radiogram has just the single audio channel and speaker system (though some designs employ more than one speaker unit, the mono signal being applied to them all in suitable frequency division). Whether mono or stereo, the f.m. front-end is a fairly common factor, the demodulator these days usually being the ratio detector in unbalanced or (in better quality equipment) balanced mode.

However, while the f.m. detector in a mono receiver feeds the audio channel direct (via the $50\mu S$ deemphasis), the f.m. detector in a stereo receiver delivers the multiplex signal to the stereo decoder, the function of which is to process the sidebands of the stereo information (the difference signal) in conjunction with the mono information and the 38kHz reference signal (derived by frequency multiplication from the 19kHz pilot tone) to retrieve the left and right transmitted audio signals in the best isolation (e.g., with the least crosstalk).

De-emphasis and sometimes subchannel suppression (deleting the residual 19kHz and 38kHz multiplex components which can interfere with the audio section and add whistles to tape recordings) are engineered into the left and right audio channels, prior to the power amplifiers.

I shall be having something to say about f.m. front-ends, i.f. channels and stereo decoders later, but for now let us concentrate on the audio channels.

The simplest audio channel is designed round the triode-pentode valve, the triode receiving the detector's signal, amplifying it and then passing it to the control grid of the pentode for power amplification, Fig.1. Quite a lot of valve radiograms are equipped with this kind of audio section, as also are the inexpensive species of record reproducers. These were in fashion when stereo started, so early stereo models boast two such audio sections.

The advent of stereo was concurrent with the development of transistor equipment, which was fortunate for the receiver designers, so the audio channels we now find in stereo eqipument are mostly transistorised.

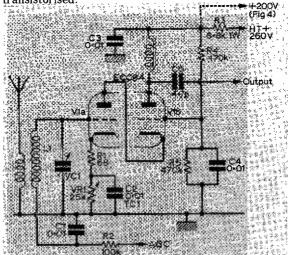


Fig. 1: A typical mono audio amplifier using a single triode-pentode valve.

Some designers favoured the use of a triode voltage amplifier (often one section of a double-triode valve) followed by a separate power pentode. This allows for greater audio power than generally possible from the pentode section of a triode-pentode valve, particularly so far as the early triode-pentode valves are concerned. Such a circuit is shown in Fig. 2, which will be seen to have much in common with that in Fig. 1.

Better quality receivers have been made with a pair of triode-pentodes in each channel to provide push-pull working. In this type of circuit the two pentodes are the push-pull output valves, while the two triodes are arranged as phase splitters supplying the anti-phase drives to the pentodes. However, since few receivers like this are in the field, there is not much point in detailing the circuit, but it is worth bearing in mind.

Let us return to Fig. 1. The audio signal from pickup or radio detector is coupled by C1 to the top of the volume control VR1, the selected level of signal then being applied via C2 to the triode grid. Note that the grid is returned to chassis through R1 (10M Ω) and that the cathode of the triode is returned direct to chassis. This is a common trick to secure grid bias for the triode. Contact potential in the valve causes a minute current to flow through R1, and because of the high value of this resistor a small

potential developed across it, and since it is negative at the grid it serves as grid bias. It requires an electronic testmeter to measure this bias (since a normal meter would swamp the $10M\Omega$ with its own resistance).

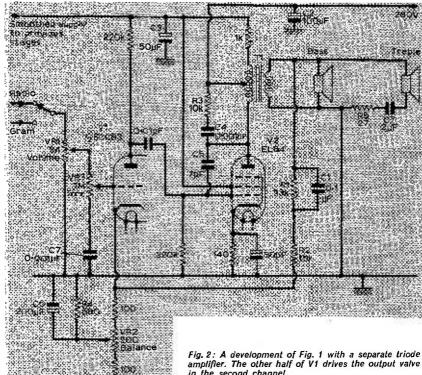
A common cause of distortion in this kind of circuit stems from a faulty triode, the fault resulting in excessive grid current and hence a potential across R1 in opposite polarity to that of the bias. The triode anode current rises, and often the valve tends to overheat.

R2 is the triode anode load, with R3 and C3 the decoupling components. The signal across R2 is coupled by C4 to g1 of the pentode, which is biased by the voltage drop across the cathode resistor R4. the grid being returned to chassis through R5. C5 is the cathode bypass capacitor (always an electrolytic in audio output stages).

The anode of the pentode is loaded by the low impedance speaker via the output transformer T1. C6 across the primary attentuates high-order harmonic distortion and makes the reproduction more palatable, as there is no negative feedback and no cancellation of second-harmonic distortion by any push-pull action.

FAULTS

Owing to the simplicity of this circuit, (Fig. 1) faults are not generally very difficult to define or locate. A cause of bad distortion is poor insulation of the coupling capacitor C4. This would permit a positive potential to exist at the control grid of the pentode with a consequent counteracting of the negative bias. It would be accompanied by excessive positive voltage at the cathode of the pentode, which could result in the failure of C5. The valve would also



in the second channel.

run very hot, possibly with the pentode anode glowing red.

Complete failure (e.g., audio drive at the top of the volume control VR1, but no speaker output) should first lead to a valve test, followed, if necessary, by voltage checks on the appropriate electrodes. C6 is subjected to audio voltage peaks which sometimes result in it shorting. This would cut the output completely, of course, but without affecting the pentode anode voltage much, so look out for this if the valve and electrode voltages seem to be normal.

Open-circuit of T1 primary deletes the pentode anode voltage, a fault condition which is often revealed by the screen grid (g2) of the pentode glowing red hot.

Low gain or volume at maximum volume control setting, assuming normal audio drive from the detector or gram pickup, is not uncommonly caused by C5 going open-circuit or low in value. This results in negative current feedback from the audio then developed across R4. The quickest way of checking this is by shunting an electrolytic across C5.

The symptom can also be caused (possibly accompanied by distortion) by the triode anode load R2 going high. If R3 goes high or C3 becomes electrically 'leaky,' the voltage at the top of R2, and hence the triode anode voltage, will drop, which again will cause low gain and possible distortion.

Before going on to more modern transistor circuits, let us have a look at some possible fault conditions in the valve circuit in Fig. 2. The faults described for Fig. 1 circuit would also be applicable here, with one or two qualifications. First, there are two separate valves involved, the triode voltage amplifier in this case being one section of a doubletriode, the other triode serving as the voltage amplifier of the partnering stereo channel (not shown).

If one channel only is in trouble one can easily compare the voltages, etc. with those at the same points in the active channel, which is a quick way of bringing the fault to light. It is also possible to substitute the same type of valves (V2, for example) between the two channels, which would tell conclusively whether the valve is responsible.

There are refinements in Fig. 2, not found in Fig. 1 circuit. For example, voltage negative feedback is applied from the secondary of the output transformer T1 to the cathode of the voltage amplifier V1, and for this to happen one side of the secondary has to be connected to chassis. The amount of feedback is determined by R1 and R2, while C1 in parallel with R1 corrects the phase of the feedback signal, ensuring that it remains substantially negative over the audio spectrum. Thus, if there is a significant rise in gain (e.g., full output at a low setting of the volume control VR1) and abnormally high distortion, then R1 and R2 should be checked for value.

C1 open-circuit could encourage instability at certain frequencies, and this is sometimes 'triggered' by transients of the music. The oscillation could be at a frequency above audio, and the only way to prove this is by connecting an oscilloscope across the secondary of the output transformer T1, preferably when the amplifier is driven by a sinewave signal from an audio generator. The oscillogram in Fig. 3 gives some idea of what the spurious signal could look like on the sinewave. The audio generator would be coupled to the grid of the triode, of course.

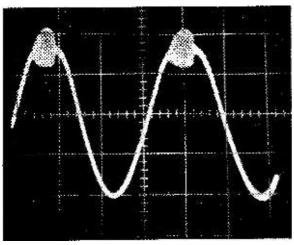


Fig. 3: This oscillogram shows spurious oscillation triggered off by transients in the signal

If the earth and feedback connections are reversed on the transformer the feedback would be positive and the amplifier would turn into a violent oscillator! Take care, therefore, when fitting a replacement transformer! Moreover, designs like this are geared to an output transformer of specific characteristics, so if it becomes necessary to replace the transformer, and the exact type is not available, then it might be necessary to alter the feedback resistors and the phase-correcting capacitor so that the values provide the correct conditions of feedback.

An interesting feature of the circuit is the tap on T1 primary. An output transformer of this kind serves both for speaker coupling and for 'smoothing' the h.t. supply fed to the earlier stages. In fact, it acts in part rather like a smoothing choke. In the

circuit, C2 is the reservoir capacitor across the power supply output and C3 is the main smoothing electrolytic.

R3 and C4 series combination across the signal part of T1 primary attenuates the higher-frequency distortion components while, with C5, providing a degree of corrective phase shift.

The triode V1 in this circuit is cathode-biased, the main cathode resistor being R4, decoupled by C6. VR2 is the stereo balance control, which feeds either side the 100-ohm resistors (one to the cathode circuit of the partnering channel). The 100-ohm resistors provide a small degree of negative current feedback since they are out of the decoupling influence of C6. The voltage developed either side of VR2 relative to the slider also contributes to the feedback. Thus it will be appreciated that by adjusting VR2 from its centre setting, the feedback on one channel will decrease, while increasing on the other channel. This, therefore, allows the gains of the two channels to be adjusted relatively for the most desirable stereo effect.

VR3 is a simple 'tone' control which works as a reactive potential-divider in conjunction with C7. With increasing frequency the reactance of C7 falls, which means that as the slider of VR3 is turned towards this end of the track the bass is lifted and the treble diminished.

It will also be seen that two speaker units are employed. That for bass is connected directly across the secondary of T1, while the treble unit receives drive via R5 and C7. R5 reduces the power, while C7 passes only the higher frequency signals. This is a very simple frequency-divider, which cannot be regarded as 'hi-fi.'

HI-FI CIRCUIT

One channel of a slightly different circuit which, possibly differing in detail, is found in some hi-fi equipment. This is the power amplifier department only, Fig. 4, where Tr1 is the input transistor, coupling to the d.c.-coupled pair Tr2/Tr3 via C1. The coupling is from Tr1 emitter (the collector 'earthed' to signal through C2). The stage is thus an emitter-follower, which has a relatively high base input impedance, giving the correct match from the control section.

Tr4/Tr5 constitute a complementary driver pair (complementary because one is n.p.n and the other p.n.p.). These drive the push-pull output transistors Tr6/Tr7 alternatively into conduction and cut-off on each half-cycle of signal. It will be noted that the base of Tr4 is fed from the same source (e.g., Tr3 collector) as the base of Tr5. Thus on a positive half-cycle Tr4 conducts and pushes a positive half-cycle into Tr6 base. At the same time Tr5 is driven downwards and since Tr7 base is connected this time to Tr5 collector a negative half-cycle appears at Tr7 base, pushing this transistor into cut-off. On a negative half-cycle of drive the conditions reverse, so that Tr6 is driven into cut-off and Tr7 into conduction.

The two half-cycles are reconstituted as a complete waveform in the load comprising the speaker, coupled through C3.

C4 is a bootstrapping capacitor which applies feedback in such a way as to increase the driver input impedance. A.C. negative voltage feedback is applied to Tr2 emitter circuit from the push-pull pair, via C5

Fig. 4: Circuit of one channel of a transistorised hi-fi stereo amplifier.

and R1. Feedback of 60dB is a common hi-fi value, which reduces the output impedance to a fraction of one ohm.

D.C. coupling exists right through to the output transistors from Tr2, so by adjusting the d.c. conditions, both the drive 'balancing' and the quiescent current of the output transistors are affected. Some circuits have no means of adjusting these parameters, so if the value of a critical component alters, so might the quiescent current and balancing.

However, in Fig. 4 VR1 adjusts the balancing and VR2 the quiescent current, which leads us neatly to hi-fi equipment fault-finding and tests. But before going on to this it is noteworthy that temperature stabilisation is often provided in or around VR2 circuit to hold the quiescent current steady with

increase in temperature of the transistors. The 'thermal element,' TH1 in Fig. 4 provides this function.

AMPLIFIER TESTS AND FAULT FINDING

For detailed testing in audio circuits, especially high quality circuits in which transistors are employed, we require an audio signal generator, audio millivoltmeter and an oscilloscope, at least. For distortion tests, we require in addition a distortion factor meter or similar instrument.

If it is assumed that the radio or pickup source signal is reaching the input of the amplifier and there is no output from the speaker, then we should check as far as possible into the amplifier with a signal tracer (e.g., pair of headphones, etc.—see Part 3, PW July 1971). In Fig. 4, for example, we could get to the base of Tr2, but possibly not any further owing to the d.c. couplings of the subsequent stages. Indeed, it is often necessary to consider the power amplifier proper as a complete circuit from the testing point of view. This is because a faulty transistor or component in one stage could reflect fault conditions into the other stages as a result of the d.c. couplings. This possibility is often overlooked when testing in circuits of this kind.

When we reach the point where the signal disappears, therefore, we may have to continue the exercise in terms of voltage measurements at the electrodes of the various transistors, comparing them with the correct values given in the service manual or on the circuit. Complete failure commonly implies significant change in the d.c. conditions, so the fault area should soon be revealed by voltage testing. If a coupling capacitor has failed, then signal tracing should reveal this without much trouble.

More difficult faults to locate are those resulting in low gain, intermittent crackling and distortion at all powers or only at high power. Starting first with distortion, a preliminary test should be that of quiescent current (assuming the amplifier to be 'Class B.' which it usually is in these days of transistors).

If possible, one should avoid connecting a milliammeter in series with the output pair for this test, owing to the resistance of the meter movement. In a circuit like that in Fig 4, the voltage across the emitter resistor (R2) of Tr7 should be measured on a d.c. millivoltmeter. The voltage will be very low because R2 is only 1 ohm! Anyway, suppose we measure 25mV, then by Ohm's law be can calculate that the current is 25mA (1 ohm resistor, since I=V/R, where I is in mA when V is in mV). A current of 25mA is, in fact, a common quiescent current value, and it can be adjusted accurately by a preset such as VR2.

If there is no preset and the current is way out, the values of the resistors associated with this part of the circuit should be measured (also the thermistor, etc., if used). Excessive quiescent current will cause the output transistors to run hot and limit the output

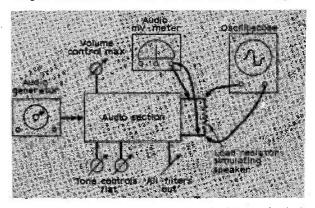
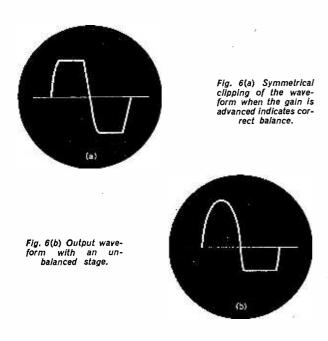


Fig. 5: Equipment set-up required to check the balance of output transistors.

power, while too low a value will precipitate high distortion (crossover distortion) at low volume.

Balancing requires an oscilloscope to monitor the output sinewave fed in from an audio generator. The setup is shown in Fig. 5 where is will be seen that the speaker is replaced by a load resistor of suitable power rating (to match the steady-state power of the amplifier). The volume control should be fully advanced, and the generator signal increased until the tips of the displayed waveform commence to clip. When the balancing is correct both tips will clip together. If one clips before the other then the balancing preset (such as VR1 in Fig. 4) should be adjusted for symmetrical clipping. The waveforms are shown in Fig. 6.



We can use the setup in Fig. 5 to measure the steady-state power of the amplifier. For this test the oscilloscope should either be accurately calibrated in peak-to-peak voltage (so that the r.m.s. voltage of the waveform can be determined by multiplying the peak voltage—i.e., half p-p voltage-by 0·7) or an r.m.s.-reading audio millivoltmeter should also be connected across the load, as shown in Fig. 5.

The load voltage just prior to the clipping point should be noted, and the steady-state power in watts is equal to the load voltage squared divided by the load value in ohms (i.e., $W=V^2/R$). The power is usually taken at 1kHz, but the power bandwidth can be appraised by noting the reduction in power at the low and high frequency ends of the spectrum.

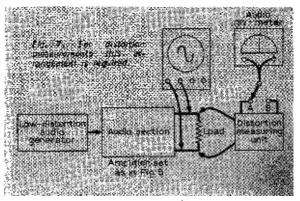
Low power, but good waveshape often signifies low supply rail voltage, and this can result from low mains voltage or incorrect mains voltage tapping.

DISTORTION TEST

Finally, total harmonic distortion testing. For this we require either a complete distortion factor meter or the set-up in Fig. 5 with a distortion measuring unit which, as was noted in the Test Instrument supplement, April 1971, is a 'notch filter' which

tunes out the fundamental frequency of the output signal, leaving only the harmonic components, the voltage sum of which is compared in ratio with the voltage of the fundamental frequency across the load, to give the dB or percentage total harmonic distortion.

The complete set-up is shown in Fig. 7. The voltage across the load is first measured at the required power (use the oscilloscope for maximum power test to indicate the clipping level) with the switch in postion A, then switching to position B, allowing the audio millivoltmeter to read the voltage of the summed harmonics. For this to happen, however, the distortion measuring unit needs to be very accurately tuned and balanced for the least indication on the audio millivoltmeter, which requires the range of the millivoltmeter progressively to be stepped down. The total harmonic distortion is the voltage ratio between the readings at switch postions A and B.



For example, if voltage A is 9V (which across 8 ohms would correspond to about 10W) and voltage B is 90mV, the ratio would be 100:1 or 40dB, corresponding to 1% t.h.d. With 9mV at B it would represent a ratio of 1000:1 or 60dB, corresponding to 0.1% t.h.d.

Remember, though, that a simple distortion measuring set 'notches out' only the fundamental, leaving wideband noise as well as t.h.d. components, so the distortion readout will include noise. Residual hum is generally removed by a high-pass filter in the test set-up, but such filtering, of course, cannot be used for distortion measurements at low-frequency for obvious reasons.

END OF PART SIX

Back Numbers Important Announcement

We regret to inform readers that owing to the closure by the Company of the department concerned it will no longer be possible to supply back numbers of **Practical Wireless** and **Television**.

To ensure obtaining regular copies of these magazines readers are strongly urged to place a regular order with their local newsagent, or to take out an annual postal subscription.

Reference to past issues of the magazines may sometimes be obtained at certain public libraries who may hold bound volumes. A few libraries are said to offer a photostat service. Alternatively, we are always willing to insert a free request for specific back numbers in our "CQ" column which appears in most issues.

practically wireless HENRY commentary by

Ted and Lid

not taking the lid off political chicanery—leave that to out-of-office politicians. TED, or, more properly, T.E.D., is a beautiful phrase I heard during a broadcast interview recently. That ubiquitous character, A. Spokesman, was defending the Post Office in the face of politic but relentless attacks from the BBC, who were following up listeners' complaints about crossed lines.

'How does it happen,' came the suave question, 'that we pick up our telephone and find our conversation interrupted by a pair of drooling lovers or an irate customer who has lost his laundry?'

Ah, well, our indefatigable spokesman countered, that is probably caused by a Transient Equipment Defect. Isn't that lovely? Imagine, Joe, when wifey grumbles about the birdies that chirp unwantedly during her 'Housewives' Choice', being able to say airily: "Can't do much about that, I'm afraid. That's TED."

The point is that the faults are not just short-lived, ephemeral, unidentified phenomena—giving them a name has fixed them, preserved them for posterity, entered them on the list of recognisable bugbears.

Wireless work is full of such



We cheer when we get a half-scale reading . . .

instances; for example, the intermittently blowing fuse. We pursue diligent tests before replacing the offending glass tube. We measure resistance, cheer when we get a half-scale reading, then remember that we intended to change the battery of the meter last time we used it.

Baffled by the lack of short-circuits we decide to beat the brute by checking the current. Meter in series—ah, which way round? Switch on. No, the other way around: is that pointer bent? But the reading is no greater than expected, and certainly less than the rating of the fuse, which we refit. Perhaps it is the result of overheating. We cover the now smug-looking set with our jacket and wait.

It is a bad habit of some manufacturers to let dirty great heatsinks jut out at the back, and the fallacy about cool-running transistors is shattered as we smell singeing Harris Tweed and realise we have a Class A output to contend with.

Touching the fins to confirm our nose, we snatch fingers back with a howl and, sucking them, muse that the problem could not have been temperature at all. And the fuse still holds.

Obviously, TED has been at work. He is only waiting for us to settle comfortably with a book and a glass at our elbow before blowing the fuse again. Or, more likely, lulling us into a false sense of security, will allow us the whole evening's listening, then, while we sleep through the small CQ hours, will sneakily contract that vulnerable thread of wire again so that rising betimes, we gaily throw the switch and get: silence!

TED is no fool. That Post Office bloke urged 'sweet reasonableness' upon subscribers when their lines were crossed. The inference was that if we pretended not to notice the fault it would go away. Any PW reader could have told him that is the wrong approach.



...and we snatch our fingers back with a howl.

How many times have you had a piece of equipment propped up on the bench, festooned with test leads, bedizened with warning lights, burbling away in the corner of your mind, and leaped into the breach as the expected fault occurs, only to be greeted with a return to normality the moment you touch the chassis?

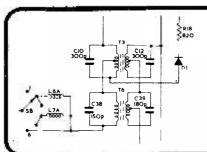
Tracking down TED is the work of an expert. One tests piecemeal. One uses not sweet reasonableness, but bitter reason. Diagnosis, as our soon-to-finish series on Servicing has emphasised, is the secret of success. And for diagnosis, where TED is concerned, read inspired guesswork. This is where LID comes in. LID is one of those contractions I am no longer likely to forget, since talking to Bob Stockwell, publicity manager of Bang & Olufsen, the day after an exhausting exhibition.

'How do you manage to organise it?'

'It's not easy,' he said, 'but we begin by catering for the Largest Imaginable Disaster and take it from there....'

That is the secret of dealing with TED. Put the LID on him: when the fuse blows, imagine the worst that could happen and assume that it has! But don't try to telephone the repairman—you will probably get the laundry company again.

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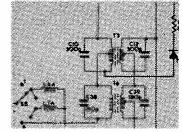


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65	2N388A 621p 2N614 30p	2N3702 17±p 2N3704 22±p	ACY18 25p ACY19 25p	AF127 171p AF139 871p AF178 45p	BC184 221p BC187 281p	BFY52 221p BSX21 371n		
)·45)·40	2N697 20p	2N3705 20p	ACY20 25n	AF178 . 45p	BC187 28ip BC213L 26ip	OC25 50p		
-45	2N706 191n	2N3711 20p 2N3819 35	ACY21 25p ACY22 20p	AF179 40p	BCY32 371p	OC26 821p OC28 621p		
·90 ·93	2N706A 121p 2N930 271p	2N3826 30p 2N3905 37p	ACY28 20p ACY40 20p	AF181 424p	BUY 10 200	OC29 75n		
-70 -50	2N1132 301n	2N3914 P.A.	ACY41 25p	AF186 664p AF289 424p ASY28 28p	BD115 78p BD121 65p	OC35 40p OC36 621p		
-60	2N1303 17±p 2N1305 25±p	R.C.A. 40253 P.A.	ACY44 40p ADI40 40p	ASY28 28p BC107 15p	BD123 821p BD124 621p	OC42 25n		
-85 -70	2N1306 25p 2N1307 25p	40398 P.A.	AD142 58p	BC108 15n	BD124 621p BD131 971p	OC45 121p		
85	2N1711 25m	2N4061 221p		BC109 15p BC113 27p	RF115 26n	OC46 15p OC70 15p		
-55 -50	2N2147 721p 2N2160 571p	3 2NA062 201n	AD161 37 p AD162 37 p	BC114 371p BC115 321p	BF117 471p	OC71 124n		
45 45	2N2614 80p	2N4286 171p 2N4291 171p AC107 30p	Ar 102 38p	BC116 624p	BF160 P.A. BF162 P.A.	OC72 121p OC74 821p		
·40	2N2646 571p 2N2905 40p	ACIII 600	AF106 421p AF114 25p	BC116A 371p BC117 39p	BF163 35p BF167 25p	OC/0 224m		
·80 ·80	2N2926 Green 14p	AC126 20p AC127 25p	AF115 80p	BC118 324n	RF173 324p	OC77 271p		
90 80	Vellow 1915	AC128 . 20p	AF116 25p AF117 25p	BC134 571p BC135 P.A.	BF178 35p BF179 721p	OC78 25p OC81 20p		
80	Orange 121p 2N3053 271p	AC154 22½p AC176 25p	AF118 60p	BC136 P.A. BC137 P.A.	BF179 7240 BF180 35p BF181 3240	OC81D 20p		
50 80	2N3055 75p 2N3391 20p	AC187 621p	AF124 224p	BC138 P.A.	BF184 25p	OC84 25p		
63 35	2N3392 20p	AC188 871p ACY17 271p	AF125 20p AF126 20p	BC142 80p BC143 P.A.	BF194 22ip BF195 27ip	OC139 321p OC140 321p		
-80	Dior	DES & RECTI	CIEDO	BC147 171p	BF196 421p BF197 311p	OC170 30p		
30 38		BZY88		BC149 174p	BF198 42#p	OC171 20p OC200 821p OC202 471p		
40 00	AA119 10p	(Series) 321p	OA202 10p	BCIS7 20p	BF200 361p BF224 30p	OC202 47-p OCP71 42-p		
50	BA102 224p	1 UA47 7+n	BA144 121p BA145 20p	BC158 171p BC169B 14p	BF225 80p	P346A 25p TIS 40p		
50 85	BA114 121p BY100 221p	OA70 7ip OA79 9p	BA148 23p	BC169C 15p	BFX84 30p	115 409		
15	BV126 20n	OA81 71p	BA155 P.A. BA156 P.A.	BC171 171p BC175 271p	BFY19 33p BFY50 22½p	P.A. Price on		
25	BY127 221p	OA90 71p		BC183 221p	BFY51 22 p	application		
25			CARTR	IDGES				
50		Inc. P.T. each		Inc. P.T.		Inc. P.T.		
	ACOS CD70		X3H S/S		RONETTE	each		
80 75	GP79	68p	X3H S/S X5M S/S X5H S/S	£1·39 £1·39 £1·39	RONETTE 105 . 8/8	99n		
80 75 75 75	GP79 GP91—18c GP91—28c GP91—38c (T	63p	X3H S/S X5M S/S X5H S/S SX5W S/S	£1.89 £1.89 £1.89 £1.81	105 8/8 106 8/8 DC400 S/8	99p 99p		
80 75 75 75 45	GP91—18c GP91—28c GP91—38c (T	63p £1.05 £1.05 £1.05 £1.32	X3H . S/S X5M . S/S X5H . S/S SX5M . S/S SX5H . S/S SX5M . D/S	£1.39 £1.39 £1.39 £1.81 £1.81	RONETTE 105	99p 99p 70p 70p 21 12		
80 75 75 75 45 50 38	GP79 GP91—18c GP91—28c GP91—38c (T GP92 GP93—1 GP94—1	C8)	X3H s/s X5M s/s X5H s/s SX5M s/s SX5H s/s	£1.39 £1.39 £1.39 £1.81 £2.00	RONETTE 105	99p 99p 70p 70p 21-12 21-12 84p		
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80 75 75 45 50 38 30 75	GP79 GP91—18c GP91—28c GP91—38c (T GP92 GP93—1 GP94—1 GP94—5 GP95 GP96	63p 	X3H	£1.89 £1.39 £1.81 £1.81 £2.00 £2.00 £2.00	RONETTE	99p 99p 70p 70p 21-12 £1-12 84p		
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450ft	4	43p	2400ft	7	£1.30
210ft	3	27p	1800ft	51	£ 1·10
Long Play		-	1200ft	5	87p
1200ft	7	70p	Double Play		
850ft	51	63p	1800ft	7	£1.00
600ft	5	50p	1200ft	51	85p
Standard Play			900ft	5	70p
Length	Spool Size in.	Price	Length	Spool Size in.	Price
POLYESTER			POLYESTER		
POLYESTER			POLYESTER		

	PTY	TAPE	REELS	CASSETTES
3in			7p	Boxed in Plastic Library Packs
4in			9p	C60 58p
5in	• •		11p	С90 63р
5≩in 7in	٠.	• •	13р	С120 87р
7111			13р	P. & P. 7p on all orders.

THE BROADCAST BANDS Malcolm Connah

Frequencies in kHz • Times in GMT

MONTHLY **NEWS FOR** DX LISTENERS

THE first report this month comes from a new reporter D. A. Hairon from Jersey in the Channel Islands. The equipment used was a Codar CR70A receiver, a 50ft long wire and a 19 metre band dipole. This combination enabled him to hear the following stations:-

9480 Radio Peking in English at 0130.

9625 CBC, Canada in English at 0200.

11710 R. Nacional de Espana in Spanish at 0130.

11720 CBC, Canada in English at 0200.

11730 R. Nederland, Bonaire, English at 0200.

11745 HCJB, Quito, Ecuador, English at 0000.

11775 TWR, Bonaire in Spanish at 0215.

11810 RAI, Rome in English at 0100. 15195 R. Ankara, Turkey, Turkish at 1600.

15250 RSA, South Africa in French at 1856.

15300 HCJB, Quito, Ecuador in Spanish at 2145, and in Portuguese at 0000.

15305 SBC, Switzerland in English at 1315.

15345 Radio Kuwait in English at 1630.

15355 ABC, Australia in English at 0530

17655 R. Cairo, U.A.R. in English at 1730.

17715 ABC, Australia in French at 2315.

17810 R. Nederland, Bonaire in Spanish at 2130.

17830 R. Lebanon in English at 1830.

17825 NHK, Japan in English at 0800.

21525 WNYW, U.S.A. in English at 1800.

21640 R. Nederland, Bonaire, English at 1830.

25730 R. Norway in English at 1200.

Ray Warren of Bury St. Edmunds has used his Koyo 1661 receiver to good effect and heard the following:-

3980 Voice of America in English at 1720.

5970 R. Free Europe, Germany at 1450.

6075 WNBC, New York, news at 1630.

6085 R. Nederland in English at 0930.

9030 R. Peking in English at 2050.

9390 R. Tirana. Albania in English at 1530.

9750 SBC, Switzerland, news in English, 1540.

15300 HCJB, Quito, Ecuador, English news, 1410.

Peter Hall of Pocklington, Yorkshire, has a Meridian 10 transistor superhet and a 50 foot aerial which enabled him to hear:-

6070 R. Sofia, Bulgaria in English at 1930.

6130 HCJB, Quito, Ecuador in English at 0800.

7275 RAI, Rome in English at 1940.

9550 Radio Finland in English at 1815.

9580 BBC, Ascension Island relay at 2115.

9620 R. Belgrade, Yugoslavia at 1800.

9625 Israel B.C. in English at 2045.

9630 R. Trans Europe, Portugal at 2100.

9700 R. Sofia, Bulgaria in English at 2140.

11620 AIR, Delhi in English at 2010.

11672 R. Pakistan in English at 2030.

11755 R. Finland in English at 0930.

11915 BBC, Cyprus relay at 1645.

11965 TWR, Monaco in English at 1700.

Ross Pullen of Crawley used his Murphy A72 communications receiver and a 100 foot long-wire to hear some interesting stations:-

6015 VOA, Rhodes, music and talk at 2200.

6145 R. Nacional, Brazil in Spanish at 2245.

9510 R. Noumea, New Caledonia, French at 0910.

9605 HCJB, Quito, Ecuador, English at 0730.

9745 R. Baghdad, Iraq in English at 2015.

9775 R. Diffusion-Television Nationale Congolaise, African music at 2200.

11672 R. Pakistan, English at 2002.

11965 Rwanda relay of Deutsche Welle at 2037.

17705 R. Havana, Cuba, Spanish news at 2015.

F. Wall of Walthamstow, London, E.17, has a homebrew 5-valve superhet and a 20 foot wire aerial, this equipment enabling him to hear:-

6035 R. Warsaw with Polish news at 1620.

6070 R. Sofia, Bulgaria in English at 1930.

9912 AIR, Delhi English news at 2000.

9630 R. Sweden 'Calling DXers' at 1230.

9660 R. Trans Europe in English at 1250.

9833 R. Budapest, Hungary, requests at 1945. 11735 R. Belgrade, Yugoslavia, English at 1530.

11740 ABC, Australia at 1530.

11755 R. Finland in English at 1500.

11765 ABC, Australia, sports programme at 0830.

11955 BBC, Malaysian relay at 1815.

15325 CBC, Canada, sports news at 1200.

21460 HCJB, Quito, Ecuador sign-on in French at 2015.

21535 RSA, South Africa, English news at 0930. 21595 CBC, Canada with news at 1910.

25790 RSA, South Africa, English news at 1500.

Stephen Mathews of Hull with his Bush 4-valve domestic receiver and 90 foot loft aerial heard the following:-

9660 YVLM Radio Rumbos, Venezuela at 0320.

9833 R. Espana Independiente, clandestine at, 2230.

11765 ZYB8, R. Diffusion, Brazil at 0220.

11795 WINB, Red Lion, U.S.A. English at 2100.

11815 ZYW24, R. Brazil Central at 0045.

15260 FEBA, Seychelles testing at 1700.

17890 BED-40, Taiwan, weak at 0330.

21690 WIBS, Grenada in English at 1945.

21720 R. Ghana in English at 1445.

Reports should arrive by the 15th of the month and be addressed to the author at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.

THE AMATEUR BANDS David Gibson, G3JDG

Frequencies in kHz • Times in GMT

IGH noise levels have been the main complaint this past month with many people confessing to giving up and going off to watch television (sacrilege!). Despite these conditions there was still a large number of listeners who persisted, despite the noise, and managed to log some worthwhile DX as reward for their labours. Just goes to show, while you are busy moaning about the QRM and QRN, someone else is just getting on with it and working half the world.

News that G5ZT has received slow-scan television pictures from FG7XT in Guadeloupe makes you think. Many people would be only too pleased to hear an FG7 on c.w. **John Rimmer** (Deal) warns that VS9MF is on from Gan—a small island in the Maldive group of islands. Listen on 14 and 21MHz; sorry, no specific freqs given.

No activity reported on ten metres this month and it rather looks as though everyone has migrated to 14MHz. This latter band is an early morning/late night affair nowadays with everything dying out from mid-morning to late afternoon. On fifteen metres the same remarks apply, although occasionally South American stations appear during the daytime. For those who would like to persist on 28MHz, suggested time for a listen is from 1400 to 1600hrs.

Incidentally, this is one of the best bands on which to play with aerials since these normally work out to a very reasonable size. A quarter wave vertical is only around the 8ft. 6in. mark, exact size depending upon frequency since ten metres is one of the widest bands we have among those l.f. of 144MHz. If sunspot counts are any indication, then September would be a very good time for ten metres

A few contests this month just to get your ears in trim for the long winter evenings. September 11 and 12, WAE phone contest; 12, 3.5MHz Field Day; October 2 and 3, u.h.f./s.h.f. Field Day; 2 and 3, Oceania contest-a chance to bag a nice list of VK/ZL callsigns; 2 and 3, IARU Region 1 u.h.f./s.h.f. contest. Don't forget the Cumulative Activity on 70MHz from October 9 to December 30. This is our "use or lose" band and all support is very welcome, if not essential. You might try listening for some of the RAEN activity on four metres. The Radio Amateur Emergency Network has numerous "local" groups and the favourite band is 70MHz. How about listening out for your local group, or pérhaps even joining them? Name and address of the Secretary of RAEN is; E. Bassett, 57 Upper St Helen's Road. Hedge End, Southampton, SO3 4LG.

Logs

A solitary log for two metres arrived this month from **David Sheekay** (Ashford). Gear is a superregen using three transistors and an i.c. audio amplifier, homebrew, and antenna is described as "telescopic". Stations logged include; G2JF, G2VB, G3EFX/P, G3JXN, G3LTF, G3TDP/M, G3UNT,

G4AAR, G4AAZ/M, G5OX, G8BYC, G8CCO, G8EBV, G8ELW and F0UJ. What about it all you converter and superhet types—you can't let a super-regen show you up, or can you?

Richard Mortimore (Cardiff), is now up to 20 w.p.m. at c.w. and says that he recommends all s.w.l.s to learn morse. Small sample from his list of c.w. pickings on 21MHz includes; CR6ID, JA3IL, KZ5BB, PY2EYE, PY4UK, W4NZK, WB4QVQ.

Mark Marsden (Ilford) uses a cassette tape recorder which has its input wired in parallel with the headphones. He claims this is very useful for checking calls heard against calls logged afterwards. Seems a very good idea. Calls heard, checked, recorded and logged on 14MHz include signals from; CE5ID, JA3JH, KV4HE, KZ5JF, PY2ZAC, PY4AEB, SV0WZ, VK3ART, VK4PJ, VP2AA, WA1DYR/MM, WB6YIY/MM, YV5CPA, YN2SC, ZL5ISJ, 4X4JU, 4X4OC, 9Q5IA.

"My uncle has given me all the PWs he can spare", writes **Kevin Davies**. Receiver is an R107 and the antenna 120ft. l.w. The following were in evidence on 14MHz s.s.b.; CT1ZE, EA3KO, EA4LM, EL2C, EL4A, HR3VSJ, JA1VK, K1WPS/CT1, LU4DMG, W1CMG/P1, W1MYA/MM, WA5KHM, WB6YIY, 4U1ITU, 4X4HT.

Paul Newman is at a new QTH which is some 680ft. a.s.l. Antenna at present is an 80ft. end fed and the receiver a modified BC348R ex G3HXS. With this gear, and from his lofty perch at Aylesbury, Paul bagged these on 14MHz; OD5BV, PY6AT, PY8LM, (Sao Luis Island), WA5STY/P/YV5 (Caracas) 4X4BL, 4X4GV, 5Z4KL, 9K2AL.

Jim Martin says that as it's the start of his holidays he thought he'd send in a report. (May the sun shine on your antenna, Sir.) His recently-purchased HA500 is nourished with an inverted V and rewards on 14MHz were; CR6EF, CR7IK, EL2CY, FP8CS, HC6CB, JA3IXL, JA4EDY, JA6FUV, JY9XL, KG4AN, KH6FP, KH6BB (Kure Island), KL7DIV, KR6US, KV4AD, KZ5AO, LG5LG, MP4MBC, PY1FN, PJ2MI, PZ1AD, PZ1AW, TG9GF, VK3JS, VP2MO, VQ9R, VS5CB, VS9MT, VU2BEZ, YV5EC, ZE1DO, ZP5GU.

On 3.5MHz c.w. it was UA1YUZ. Up on 7MHz, signals came from CX6CG, LX1BJ, PYCVX, PZ1AX, UW3NG, ZB2A, and ZP5AR. Gear is a JR500SE with 132ft. end fed. Operator's ears belong to **Julian Iredale** (Llandudno), and up on 14MHz s.s.b. Julian heard CR6GA, HV3SJ, OG5A, (Finnish Field Day prefix), PY2ZAC, PY4ATL, PZ1DL, TI2J, 5A4DL.

P. McKay advises that he never notices any logs for 7MHz in the Amateur Bands, and sends in a list of G-stations heard on this band. Well, the G's must have been working someone, so how about a listen this month?

Logs, in alphabetical order please, to arrive by the 15th of the month to:

12, Cross Way, Harpenden, Herts.



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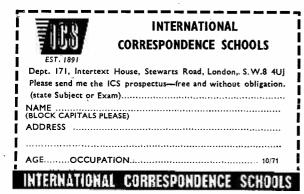
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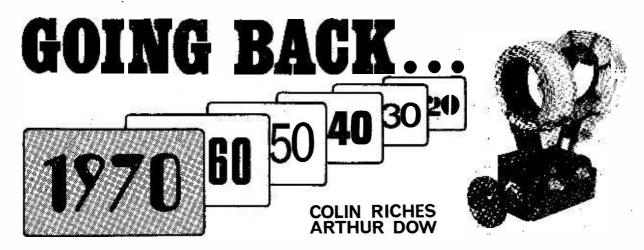
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ANY readers have written to ask us if we can suggest any gramophone records that may help them to bring alive even more vividly the days of vintage radio. After a certain amount of research we have discovered that Fountain Press, the book publishers, have now announced a new "Fountain" record label of Vintage Jazz music. The first 12in. LP releases feature historic jazz and blues recordings but future plans include Music Hall, Opera and other music of the "Good Old Days".

The records are housed in attractive double sleeves which open, displaying an authentically documented note section with accurate historical background information and photographs.

The first two releases are: "Original Dixieland Jazz Band and the Louisiana Five"—from the extremely rare Aeolian sessions of 1917 (catalogue number FJ-101). "Ladds Black Aces—Vol. 1"—the first 16 sides recorded by another pioneer white jazz band during 1921-1922 (Cat. No. FJ-102). Following soon are "Naylor's Seven Aces" (c 1923-4). "Jelly Roll Morton" 1923-4 piano solos. "Lovie Austin's Blues Serenaders" (1924-6), and "Annette Hanshawe. Vol. 1"—Miss Hanshawe enjoyed public acclaim as a singer of popular songs in the 1920's and 1930's and her records are great collectors' items today.

These records are available from record shops at £2·10 or direct at £2·25 which includes post and packing, from Fountain Records, 46 Chancery Lane, London WC2A 1JU.

If you have any questions or queries on these records or any other Vintage records or performers please write to us at Practical Wireless and we'll do our best to help you.

The Mystery of the ST.200

We would, at this point, like to thank the many enthusiasts who have been kind enough to write and help us try to clear up the mystery of the missing John Scott Taggart design "ST.200" mentioned in "Going Back", August 1971.

One theory put forward by Mr. F. W. Snow, Belper, Derby, is that there never was a design called the ST.200 but one called "The ST Twin"—a 2-valve reflex. It was published in *The Wireless Constructor* and not the *Popular Wireless* magazine during the late 1924's and early 1925's.

Mr. J. Crossley hailing from Failsworth, Manchester, also maintains that the ST.200 never existed. He

says that J.S.T. visited the United States just after the advent of the ST.100. On his return, he was full of enthusiasm about a gentleman by the name of Armstrong, and *Popular Wireless* published a circuit by this man which J.S.T. had brought back with him. It was the "Armstrong Regenerative Circuit". J.S.T. then went on to publish the ST.300 design, so it appears that Armstrong's receiver may have taken the place of the missing ST.200.

Mr. A. H. Jenkins, Dringhouses, York, writes to say that as far as he recollects, the ST.200 may well have been just a mains-operated version of the ST.100 and therefore did not differ operationally from that design. It did however, need an extra valve. Mr. Jenkins also informs us that the S.T. series did not start at ST.100, although this was the most famous. He says that he used to have a limp covered book of J.S.T. circuits which started at crystal sets and worked up through one valvers. He seems to remember an ST.37 which was quite well-known at the time.

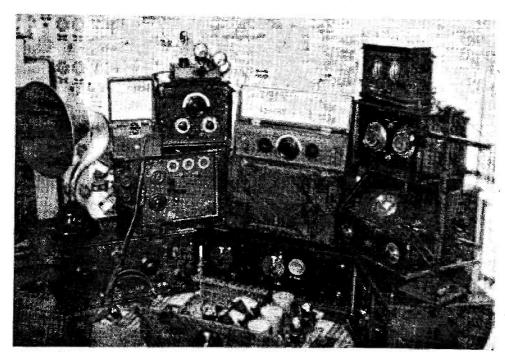
Vintage Sets

Mr. William Pozniak, from 37 Manley Road, Oldham, Lancashire, has loaned us some photographs. He says that for three years he has been routing through old cellars and derelict radio shops throughout the North (we hope you came out for food now and again, Mr. Pozniak) and has produced the collection of Vintage gear shown in the first picture. He says that it takes him most of his time to strip down and restore these sets to as near as possible their original condition. The sets he has are as follows: The Burndept Screened Four, The Burndept Portable Five, the Burndept Ethophone Five and the Burndept Ethophone Two.

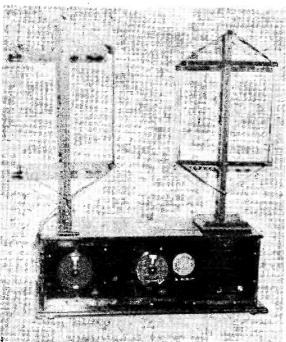
Mr. Pozniak believes there were about 20 Burndept receivers. This figure is agreed by Mr. Keith Lancaster who we mentioned some time ago as wanting to start a Vintage Radio Club. Incidentally, response has not been too good and if readers are interested in such a Club, would they please contact Keith Lancaster, 40 Great Gardens Road, Hornchurch, Essex.

The rest of Mr. Pozniak's collection contains a Marconiphone Type 32 three valver, a domestic Cossor two valver, a small set in a metal case named a GECoPHONE Victor 3 and a few home-constructed receivers

The receiver shown in the second picture is a



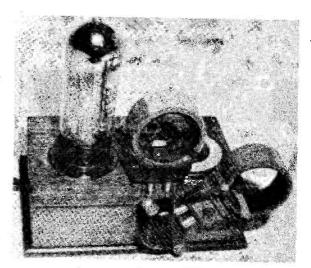
An overall view of Mr. Pozniak's collection of Burndept receivers. The small receiver shown at the top centre of the picture and the set in the centre of the picture together with frame aerials are shown enlarged below.



The James Super Six All Waver (c 1927).

1927 James Super Six All Waver (medium, long and short). The reader says that he purchased this set from an old radio shop which was a booming concern in the 1920's, together with a BTH horn speaker, and the person who sold it to him, told him, as he removed the set from the dusty shelf, that he was the first person to touch the set in 40 years.

The neat little receiver shown in the third photograph seems to Mr. Pozniak to be an unusual type, for he says that there appear to be two resistors and



Louwe receiver with encapsulated components.

two capacitors encapsulated in glass tubes which have then been assembled inside the glass envelope of the valve. The make on the case reads "Louwe Radio Berlin" and on the valve it states "Louwe Radio British Made". This set is in working order on headphones although it was originally made to drive some sort of speaker. The reader would like to know the date and origin of this set if anyone can help him and he would also be interested if anyone could supply him with original Burndept valves for any of their makes of receivers.

TITANIC DISASTER

We would like to thank the many readers who have written to us about the Titanic Disaster. We are gradually answering all the letters and we hope you will bear with us.

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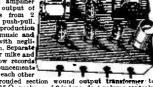
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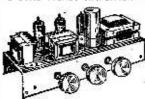
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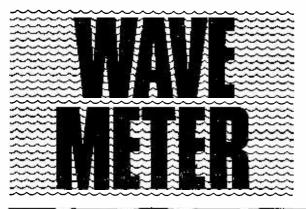
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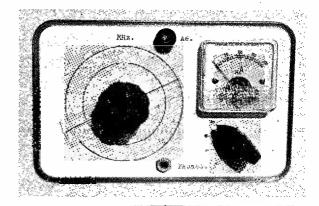
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R. F. GRAHAM

THIS unit has four switch-selected bands, covering from 1600kHz to 100MHz and it can be used as a wavemeter, field strength meter or phone transmission monitor. It is fully enclosed, selfcontained, and requires no battery.

Figure 1 shows the circuit. One pole of the switch, S1, selects the inductors L1, L2, L3 or L4, which are tuned by VC1. Pole S2 switches the diode D1 to the appropriate small coupling winding. When the circuit is tuned to resonance by VC1, rectified current is indicated by the meter.

To monitor phone signals, high impedance headphones or earpiece can be plugged into the phone jack and the phones are then in series with the meter. Cl is for r.f. and a.f. by-pass purposes, while C2 is an r.f. by-pass capacitor.

Fig. 1: The circuit of the wavemeter monitor. For the details of coil construction see text. C1 C2 0•01 Jack

INDUCTORS

L1, L2 and L3 are wound on 12in. diameter paxolin formers. These can be prepared by cutting 112in. lengths of ¹2in. diameter tube, and drilling small holes to secure the ends of the windings. The tubes can later be fixed with adhesive, with small brackets or by having them a push fit on small discs of insulating material screwed in place. Surplus or junk box formers could probably be utilised, but if these are of different diameter it may be necessary to modify the number of turns to suit.

L1 is wound with 34 s.w.g. enamelled copper wire. The tuned portion A to B is 65 turns, in a compact pile about ³16in. long. A space of about ³16in. is left and section B to C is wound on in the same direction, and has 8 turns side by side.

L2 also uses 34 s.w.g. enamelled copper wire. A to B is 50 turns wound side by side. A space is left as before, and a 4 turn coupling winding ends at C.

L3 uses 24 s.w.g. enamelled wire and has 1212 turns from A to B, and 2 turns from B to C. Figure 2 shows the general construction of L1, L2 and L3.

The windings are doped, smeared with clear

Bostik 1 or with some other suitable coil cement. L4 is self-supporting, of 16 s.w.g. bare wire, and is shown in Fig. 3. It is wound and shaped to have 8 turns in all and is 12in. outside diameter, and 1in. long. B is a tapping 2 turns from end C. This coil is soldered directly to tags A and C of the 4-way switch as shown in Fig. 3. A lead runs from tapping B

★ components list

200pF Wavemaster variable capacitor. C1 0.04 fel or singlar. 0 01gf or similer. OA30 or OA91 diode. C2

S1759 T-pole & way miniature rotary switch.

Small closed dutillt lack. 200p. A ministore meter

Industris as led

Priceting persons, box about 6 x 3 x 2 in., aerial socket, andhs etc

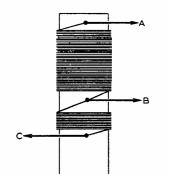


Fig. 2: The general construction of coils L1, L2 and L3 wound on ½in. diameter paxolin formers. Details of L4 are given in the text and in Fig. 3

directly to the moving plates tag of the variable capacitor.

CONSTRUCTION

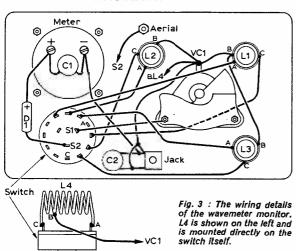
The wavemeter was built to occupy a rigid plastic 'snack box' approximately $6in.\times 3^12in.\times 2^12in$. deep. A piece of paxolin sheet, a little under $6in.\times 3^12in$., was cut to fit inside the box lid. The wavemeter is built wholly on this paxolin panel.

Screws holding L1, L2 and L3 are countersunk. A piece of stout paper to bear dial markings is cut the same size as the paxolin, and cemeted to it. The wavemeter is then assembled and calibrated and the scales are marked. The meter and fixing nuts of the switch, jack, and VC1 are then removed and the paxolin panel is fitted inside the box lid. This places the scales under the transparent material of the lid. Paxolin and lid are held together by the meter and the nuts mentioned.

Holes for the jack, switch and VC1 are first drilled in the paxolin and the meter hole cut with an adjustable washer or tank cutter. The paxolin is then placed in the lid, and the positions of holes marked on the lid itself, so that they can be made to match up correctly with those in the paxolin.

The box should be of material which is reasonably strong. It should be drilled and cut with tools which are really sharp, and without forcing, to avoid cracking.

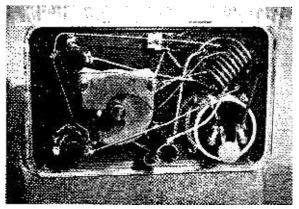
A suitable case could be assembled from ¹₄in. thick wood, glued and pinned at the joints. The paxolin panel would then be large enough to be fixed with small wood screws.



All wiring is shown in Fig. 3. Connections should be of stout wire and run clear of each other, and should be reasonably short and direct, especially for L3 and L4.

VCl is 200pF, but this exact value is not essential, provided that the ranges overlap a little. If an alternative diode is fitted, be sure it is suitable for frequencies up to 100MHz.

A $100\mu A$ meter would be a little more sensitive than the one specified, but would also cost a little more. Should a $500\mu A$ meter be to hand, it is suitable, though less sensitive. The sensitivity with a 1mA meter becomes rather low so an instrument in the neighbourhood of $200\mu A$ is probably most suitable.



An internal view of the completed prototype.

The dial cursor is a 2^{3} 4in. piece of perspex about 5 8in. wide, marked with a hair line. Readings will be confused if VC1 can rotate through 360° , so the extreme corner of the back moving plate was turned up with pliers to limit rotation to about 180° .

Values used for C1 and C2 are not particularly important. A medium or high impedance pair of headphones will give best general results.

CALIBRATION

Actual band coverage with the inductors described is as follows

Range 1 1 · 6 · 4MHz Range 2 2 · 8 · 10MHz Range 3 8 · 32MHz Range 4 21 · 100MHz

This gives 160 and 80m bands in Range 1, 80 and 40m bands in Range 2, all h.f. bands 20-10m in Range 3, and 15-10m bands and harmonics in Range 4, without having to use the extreme h.f. end of the ranges, where readings become close together.

The simplest means of calibration is to use a grid dip oscillator, placing its coil 2in. or so from the appropriate inductor in the wavemeter. This allows complete calibration of all ranges with a minimum of difficulty.

A transmitter can be used for calibration in the amateur bands, and these will probably be of most importance. Place the wavemeter some distance from the transmitter, its dummy load or the aerial lead or tuner, as required to obtain a deflection on the meter.

Some signal generators have an r.f. level meter. If the output of the generator is coupled to the wavemeter by a small loop, a dip may be observed on the level meter when the wavemeter is tuned to resonance. This offers another possible way to secure calibration.

If the wavemeter is coupled to a receiver having a sensitive S-meter, a *slight* dip in the S-meter reading can be observed when the wavemeter is tuned to the receiver frequency. Input to the receiver can be from a signal generator, or transmission.

All calibration is made with no external connection to the wavemeter and minimum coupling to avoid unnecessary inaccuracy.

USES

A wavemeter is normally used to indicate frequency bands and harmonics, not exact frequencies within a narrow band. It is placed or held to pick up sufficient r.f. energy from the circuit being investiga-

-continued on page 534

BRAND NEW GUARANTEED

SEMICONDUCTORS & COMPONENTS

SOLE AGENTS FOR SOLID STATES DEVICES INC., (USA). IN U.K.

RETURN OF POST SERVICE

						SISTO	RS				1	TTL.	LOGIC
2G301 2G302	2 201	2N3405	45p	40311	45p 35p	BCY30	27 1 p	BSX28 BSX60	32½p 82½p	NKT281 27½p NKT401 87½p	SN7400		1-11 1 20p
2G303	i 42∔n	2N 3415	22 t p	40314	47±p 87±p	BCY32	30p 50p	BSX61 BSX76	62 i p 22 i p 27 i p	NKT402 90p NKT403 75p	SN7401 SN7402		20p 20p
2G308	30p	2N3417	37 t p	40323	47 p 32 p	BCY33 BCY34	25p 30p	BSX77 BSX78	27½p 27½p 27½p	NKT404 621p NKT405 75p	SN7403 SN7404		20p 20p
2G371 2G374	l 9.0m	2NI3579	071 n	40396	47.4p 37.4p	BCY38	40p 60p	BSY10 BSY11	271p	NKT406 621p NKT451 621p	SN7405 SN7406		20p 80p
2G381 2N404	1 22∮p	2N3600	27±0 27±0	40329	30p 27}p	BCY42	50р 1 5р	BSY24 BSY25	15p 15p	NKT452 621p NKT453 471p	SN7407 SN7408	,	20p 20p
2N696 2N697	175	2N3702	221p 11p	40348	571p 521p	BCV54	15p 32էր	BSY26 BSY27	17≟p 17≟p	NKT603F321p NKT613F321p	SN7409 SN7410		20p 20p
2N698 2N706	12ip	2N3704	11p	40361	42±p 47±p	BCY59	32½p 22½p 22½p 97½p	BSY28 BSY29	17‡p 17‡p	NKT674F 30p NKT677F 30p	SN7411 SN7412		23p 48p
2N705 2N708	15n	2N2706	09p	40370	57∔p 32∤p	BCY60 BCY70	20p	BSY32 BSY36	25p 25p	NKT713 25p NKT781 30p	SN7413 SN7416		40p 84p
2N709 2N718	25p	2N 3708	11p 07p	40406 40407	57∦p 40p	BCY71 BCY72	25p 174p 274p	BSY37 BSY38	25p 22‡p	NKT10419 30p NKT10439	SN7417 SN7420		84p 20p
2N726 2N727	′ 80p	2N 3710	09p 09p	40410	52∔p 62∔p	BCZ11	424D	BSY39 BSY40	22½p 32½p	37½p NKT10519	SN7423 SN7425		51p 48p
2N914 2N916	17 p	2N3711 2N3715		40467.A 40468.A	1 35p	RDIM	65n	BSY51 BSY52	32‡p 32‡p	32½p NKT20329	SN7427 SN7428		48ր 80թ
2N918 2N929	22 1 p	2N3716 2N3791	£2.06	40600 AC 107 AC126		BD123 BD124	821 p 60 p	BSY53 BSY54 BSY56	371p 40p	47½p NKT20339	SN7430 SN7432		23p 48p
2N930 2N109 2N109	0 22 līp	2N3819 2N3823 2N3854	35p 97±p 27±p	AC127 AC128	20p 25p 20p	BD131 BD132 BDY10	75p 85p	BSY78 BSY79	90p 47½p	37½p NKT80111	SN7433 SN7437		80p 64p
2N113 2N113	1 25p 2 25p	2N3854, 2N3855	A 271p 271p	AC154	221p 25p	BDY11 BDY17	£1.624	BSY82 BSY90	45p 521p 571p	77½p NKT80112 97½p	SN7438 SN7440		64p 23p
2N130 2N130	2 17±p	2N3855. 2N3856	A 80p	AC187 AC188	62 էր	BDY18 BDY19	£1.75	BSY95A BSW41	121p 421p	NKT80113 £1.12	SN744LA SN7442	ıN	87p 85p
2N130 2N130	4 224p 5 224p	2N3856. 2N3858	A 35p	ACY17 ACY18	37 p 27 p 25 p	BDY20 BDY38	£1.12‡	BSW70 C111	271p 75p	NKT80211 924p	SN7443 SN7444		£2.86 £2 £2.86 £2
2N130 2N130	o 25p 7 25p	2N3858. 2N3859	A 30p 27½p	ACY19 ACY20	25p	BDY60 BDY61	£1.25 £1.25	C424 C425	27½p 55p	NKT80212 921p	SN7445		£2-50 £2 SILICO
2N130 2N130	8 30р 9 30р	2N3859. 2N3860	A 321p 30p	ACY21 ACY22	25p 20p	BDY62 BF115	£1.00 25p	C426 C428	40p 37≟p	NKT80213 92½p	PIV	50	100 200
2N150 2N161	7 171 p 3 25 p	2N3866 2N3877	£1.50 40n	ACY28 ACY40	20p 20p	BF117 BF163	47½p 37½p	C744 D16P1	30p 37∤p	NKT80214 924p	1A 3A	10p 15p	12½p 15p
2N163 2N163	2 30n	2N3877 2N3900	A 40p 87*p	ACY41 ACY44	40p	BF167 BF173	18p 19p	D16P2 D16P3	40p 37≟p	NKT80215 924p	6A 10A		25p 52½p 57½
2N163	9 27 1 p	2N3900. 2N3901	974 p	AD140 AD149	57 l p	BF177 BF178	30p 30p	D16P4 GET102	40p 30p	NKT80216 921p	17 A 35 A	_	521p 571 571p 621 80p 90p
2N171 2N188	1B £1·00 1 25p	2N3903 2N3904	35p 35p	AD150 AD161	621 p 371 p 371 p	BF179 BF180	30p 35p	GET113 GET114	20p	OC20 75p OC22 50p	1 amp ar		are plastic e
2N189 2N214	3 97 in	2N3905 2N3906 2N4058	371p	AD162 AF106 AF114	42½p 25p	BF181 BF184 BF185	32½p 25p	GET118 GET119 GET120	20p 20p	OC23 60p OC24 60p OC25 50p	IN34A	10p	DIODES
2N214 2N216	8 5 7≨p	2N4059 2N4060	174p 10p 124p	AF115 AF116	25 p 25 p	BF194 BF195	42½p 17½p 15p	GET873 GET880	52½ p 12½ p 30p	OC25 50p OC26 27½p OC28 62½p	IN914 IN916	07±p 07±p	AA119 AA129 AAZ13
2N219 2N219	3 40p	2N4061 2N4062	12½p 12½n	AF117 AF118	25p 62½p	BF196 BF197	42½p	GET887 GET889	20p 224p	OC29 621p OC35 50p	IN4007	22½p 10p	AAZ15 12 AAZ17 12
2N219- 2N221	4A 800	2N4244 2N4285	471p	AF119 AF124	20n	BF198 BF200	421p 521p	GET890 GET896	22½p 22½p	OC36 62½p OC41 22½p	18113	15p 15p	BA100 15 BA102 22
2N2218 2N2218	7 27½p 8 23p 9 23p	2N4286 2N4287	171p 171p 171p	AF125 AF126	22½ p 20p 20p	BF224 BF225	14p 19p	GET897 GET898	221p 221p	OC42 25p OC44 20n	IS121	17½p 12½p	BA110 32 BA115 07
2N222 2N222	0 25p 1 25p	2N4288 2N4289	17≟p 17≟p	AF127 AF139	17∄p 37∄p	BF237 BF238	23p 23p	MJ400 MJ420	£1·071 £1·121	OC46 122p OC46 15p	IS131 TS132	12}p 15n	BA141 32 BA142 32
2N2229	0 47 <u>∤</u> p	2N4290 2N4291	17½p 17½p	AF178 AF179	42½p 72½p	BF244 BFW61	23p 474p	MJ430 :	£1.12} £1-02}	OC70 15p OC71 124p	18920 18922	07≟p 07≟p	BA144 12 BA145 20
2N229 2N2368 2N2368	8 17tm	2N4292 2N4303 2N5027	121p 471p	AF180 AF181 AF239	521p 421p	BFX12 BFX13	22½p 22½p	MJ440 MJ480	95p 97₁p £1.25	OC72 121p OC74 321p OC75 221p	18923 18940	07∄p 07∄p	BA154 12 BAX13 12
2N2369 2N2416	9A 17∳p	2N5028 2N5029	521p 571p 471p	AF279 AF280	42½p 47½p 62½p	BFX29 BFX30 BFX42	30p 30p 37}p	MJ481 MJ490 MJ491	£1.25 £1.00 £1.37‡	OC75 221p OC76 221p OC77 30p		TDIA	
2N2488 2N2484	3 27±p	2N5030 2N5172	421p 121p	AF211 ASY26	321p 25p	BFX44 BFX68	37±p 67±p	MJ1800 MJE340	£2·17½ 62½p	OC81 20p OC81D 22in	SC35D £	TRIA:	SC51D £1.9
2N2539 2N2540) XX*D	2N5174 2N5175	52½p 52½p	ASY27 ASY28	37½p 27½p 27½p	BFX84 BFX85 BFX86	25p 32½p	MJE520 MJE521	60p 73p	OC83 25p OC84 25p	SC36D 2	81 00 ¦	40430 97 40486 95
2N2613 2N2614	1 30p	2N5176 2N5232A	45p	ASY29 ASY36		BFX87	25p 27}p	MPF102 MPF103	42 է P 87 է P	OC139 324p OC140 324p	SC45D £	1.621	40528 72 40430 £1.8
2N2646 2N2696 2N2711	5 32 <u>∤</u> p	2N5245 2N5246 2N5249	45p 42½p 67½p	ASY50 ASY51 ASY54	25p 32½p 25p	BFX88 BFX89 BFX93	25p 62ip 1 70p	MPF104 MPF105 MPS3638	37 ± p 37 ± p	OC170 30p OC171 30p OC200 40p	SC46D £		40432 £1.8 40512 £1.4
2N2712 2N2713	2 25p	2N5265 2N5266	£3.25 £2.75	ASY86 AU103	32½p £1.25	BFY10 BFY11	321p	NKT0013 NKT124	3 47‡p	OC200 40p OC201 60p OC202 75p	THYRI PIV 50	STORS	S 200 300 4
2N2714 2N2865	1 30թ	2N5267 2N5305	£2.62± 37±p	ASZ21 BC107	421p 10p	BFY17 BFY18	221p 321p	NKT125 NKT126	271p 271p	OC202 75p OC203 42½p OC204 42½p	1A 25	p 27½p	37½p 40p 4 57½p — 7
2N2904 2N2904	4 30p	2N5306 2N5307	40p	BC108 BC109	10p 10p	BFY19 BFY20	32 ip £1.60	NKT128 NKT135	27½p 27½p	OC205 90p OC207 75p	5A — 7A —	- 55p - 55p	65p — 9 65p — 9
2N2905 2N2905	37∔p A 40p	$2N5308 \\ 2N5309$	37}p 62}p	BC113 BC115	15p 15p	BFY21 BFY24	42½ p 45p	NKT137 NKT210	32½ p 30p	OCP71 424p ORP12 624p	Also 12 a	mp, 100	00 PIV 55 p. PIV 75 p
2N2906 2N2906	A 271p	2N5310 2N5354	421p 271p	BC116A BC118	10p	BFY25 BFY26	25p 20p	NKT211 NKT212	30p 30p	ORP61 50p	2N3525 a	it £1.12}p 30ARD	
2N2907 2N2923	15p	2N5355 2N5356	27½p 32½p	BC121 BC122	20p 20p	BFY29 BFY30	50p 50p	NKT213 NKT214	30p 22≟p	TIS34 62½p TIS43 27p		(0-15 0-1 atrix Matrix
2N2924 2N2925 2N2926	15p	2N5365 2N5366 2N5367	471 p 321 p 571 p	BC125 BC126 BC140	20p 20p 37≨p	BFY41 BFY43 BFY50	50p 62½p	NKT215 NKT216	22±p 37±p	TIS44 10p TIS45 10p TIS46 11p	2½ × 3½in 2½ × 5in 3½ × 3½in	1 3	17½p 20p 21n 24n
Green Yello	n 14p	2N5457 2S005	37‡p 75p	BC147 BC148	10p 10p	BFY51 BFY52	23p 20p 23p	NKT217 NKT219 NKT223	42 i p 30 p 27 i p	TIS47 11p	$3^3_4 \times 5$ in	:	21p 24p 27½p 27½
Oran 2N3011	ge 12½ p	28020 28102	£2.00 50n	BC149 BC152	12p 17₁p	BFY53 BFY56A	174 n	NKT224 NKT225	25p 22½p	TIS48 121p TIS49 121p TIS50 171p	5 × 17in Vero Pins	(Bag of	85p 36) 20 p
2N3014 2N3053	82≟p 18p	$28103 \\ 28104$	25p 25p	BC157 BC158	20p 11 n	BFY75 BFY76	30p 42½p	NKT229 NKT237	30p 35p	TIS51 121p TIS52 121p	Vero Cutt Pin Inser matrix)	tion Too	ols (•1 and
2N3054 2N3055	46p 62p	$28501 \\ 28502$	32}p 35p	BC159 BC160	12ր 62] թ	BFY77 BFY90	57½p 67½p	NKT238 NKT240	25p 27‡p	TIS60 22±n	HEAT		**
2N3133 2N3134	30p	28503 3N83	27½p	BC167 BC168B	11p 10p	BFW58 BFW59	27½p 25p	NKT241 NKT242	27 <u>1</u> p 20p	TIS61 25p TIS62 27½p	4.8×4	× 1in F	finned for T 4.8×2 ×
2N3135 2N3136	25n	3N128 3N140	70p 77∤p	BC168C BC169B	11p 11p	BFW60 BPX25	25p £1.85	NKT243 NKT244 NKT245	62½p 17½p	TIP29A 50p TIP30A 60p	Finned.	for One	TO-3 Tran p. For TO TO-18,
2N3390 2N3391	20p	3N141 3N142 3N142	72±p 55p	BC169C BC170	12p 12ip	BPX29 BPY10	£1.45	NKT261	20p 20p	TIP31A 62±p TIP32A 75p	5p Finn Finned. F	ed. For or TO-18,	TO-18, 1/- Finned.
2N3391 2N3392 2N3393	17±p	3N143 3N152 R.C.A.	671p 871p 521p	BC171 BC172 BC175	15p 15p 22¦p	BRY39 BSX19 BSY20	47½p 17½p 17½p	NKT262 NKT264	30p 20p 20p	TIP33A £1.021p TIP34A £2.05	RESIST	ORS	-
2N3394 2N3402	15p 22∔n	40050 40251	55p 32↓p	BC182 BC183	10p 09p	BSX20 BSX21 BSX26	372p 45p	NKT271 NKT272 NKT274	20p 20p 20p	111 94H 22.09	Carbon Fil	%, 1p.	å₩, 1₩ & 2
2N3403	22½p	40309	32 <u>∤</u> p	BC184	11p	BSX27	47½p	NKT275	20p	(MINI)	watt 59	%, 1½p. %, 2p.	E24 Serie
	iosť ox P	Matching	charge	e (audio	transist	ors only)	i 12∄p e	wealth (A	nr) 65p	(Pilly.)	1 watt 100	%, 2 p.	¼W & ↓W E12 Series
T-1	01 45					n withou			5.	Sen	2 watt 109		nprehensive
1	01-452 Tolov	2 0161/ 21402	2/3	A. 1	MIT	RSH	AL	L&	SC		ductor and	I.C. list ((24 pages).

	TTL. LOC	SIC I.C	. NEW	PRICES	
O	1-11	12-24	l	1-11	12-24
SN7400	20p	18p	SN7446	£1.00	95 p
SN7401	20p	18p	SN7447	£1.00	95p
SN7402	20p	18p	SN7448	£1.00	95p
SN7403	20p	18p	SN7449	£1.00	95p
SN7404	20p	18p	SN7450	20p	18p
SN7405 SN7406	20p	18p	SN7451	20p	18p
	800	75p	SN7453	20p	18p
SN7407	20p	18p	SN7454	20р	18p
SN7408	20p	18p	SN7460	20p	18p
SN7409	20p	18p	SN7470	40p	38p
SN7410	20p	18p	SN7472	32p	30p
SN7411	23p	21p	SN7473	43p	41 p
SN7412	48p	46p	SN7474	43p	41p
SN7413	40p	38p	SN7475	45p	44p
SN7416	84p	78p	SN7476	45p	44p
SN7417	84p	78p	SN7480	70p	65p
SN7420	20p	18p	SN7481	£1.40	£1.38
SN7423	51p	47p	SN7482	87p	82p
SN7425	48p	45p	SN7483	87p	82p
SN7427	48p	45p	SN7484	£2·00	£1.85
SN7428	80p	75p	SN7485	£3.62	£3.40
SN7430	23p	21p	SN7486	83p	30р
SN7432	48p	45p	SN74174	£2.40	£2.30
SN7433	80p	75p	SN74175	£1.68	£1.60
SN7437	64p	60p	SN74176	£2·64	£2 55
SN7438	64p	60p	SN74177	£2.64	£2.55
SN7440	23p	21p	SN74180	£2·13 .	£2.05
SN744LAN		83p	SN74181	£9.33	£9-00
SN7442	85p	81p	SN74182	£2.03	£1.95
SN7443	£2.86	£2.70	SN74184	£4·80	£4.60
SN7444	£2·86	£2·70	SN74185	£4·80	£4.60
SN7445	£2.50	£2.40	SN74190	£1.80	£1.70
	SILIC	ON R	ECTIFIE	RS	

50	100	200	400	600	800	
n	701-	15-	10-	771 -	10	

PIV	50	100	200	400	600	800	1000	1200
1.A.	10p	12½p	15p	16p	17}p	19p	20p	
3A.	15p			22 ł p		30p		
6A.			25p	30p	$32\frac{1}{2}p$	35p	_	
10A		521p	57åp	65p	771p	86½p	97½p	£1.25
17A		57½p	62 fp	77 tp	900	97 åp	£1.20	£1.574
35A	_	80p	90p	£1.00	£1.25	£1.50	£2.50	wx-019
	and 3 amp		stic end	apsulati	on.	W. 00	U	

S & RECTIFIERS

IN34A	10p	AA119	10p	BAX16	12 1 p	BYZ13	25p
IN914	07 l p	AA129	10 _D	BAY18	17 a D	FST3/4	22½p
IN916	07 lp	AAZ13	10p	BAY31	07 P	OA5	17åp
IN4007	221p	AAZ15	12½p	BAY38	25p	OA10	22 i p
IS44	10p	AAZ17	12½p	BY100	17 åp	OA9	10p
IS113	15p	BA100	15p	BY103	22 i o	0A47	07 i p
IS120	15p	BA102	22±p	BY122	471p	OA70	07 p
IS121	17}p	BA110	$32\frac{1}{2}p$	BY124	15p	0A73	10p
IS130	12½p	BA115	07½p	BY126	15p	OA79	09p
IS131	12½p	BA141	32½p	BY127	17 a	OA81	07 lp
IS132	15p	BA142	32 ½ p	BY164	57+p	OA85	07ip
18920	07½p	BA144	12½p	BYX10	22 i p	OA90	07åp
18922	07 <u></u> ₽p	BA145	20p	BYZ10	35p	OA91	07 î p
18923	07 <u>ł</u> p	BA154	12½p	BYZ11	32½p	OA95	07 1 p
IS940	07 <u>k</u> p	BAX13	12}p	BYZ12	30p	OA200	10p
						OA202	10p

10010	0120	DAXI	0 122p	DIZZ	оор	OA200 OA202	10p		
	TRI	ACS		BRIDGE RECTIFIERS					
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4A	47 p		57±p		77 tp
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"HAT's an automatic light?" you may well ask. Well, lights are only needed when it's dark, and a very simple electronic switch circuit can be made to turn a small light on as soon as the general light level drops below a certain level; this has its most obvious application in motor cars. Many parts of the country have local regulations requiring cars to display lights shortly after sunset, even when parked. In other areas car parking lights are needed only when the local street lighting is turned off, which in some areas is as late as 2 a.m. There cannot be many people who relish the thought of going out to their cars at that hour to turn on the side or parking lights.

Porch lights which automatically turn themselves on after dusk are another application and although the circuit as shown is not really meant for this, many readers will realise the potential and modify the circuit to operate a thyristor or relay to apply the mains supply to the porch light.

THE CIRCUIT

Like last month we are making use of a light dependent resistor, that is a resistor whose value in ohms falls with an increase in light falling on it. Also like last month this component is coupled into the base bias potential divider circuit, thus controlling the bias on the transistor, Trl. R1 may have to be altered in value in order to make the "switch" operate at the correct light level.

As darkness falls, the potential at the base of Tr1 falls as the value of the LDR rises and after a certain point Tr1 begins to turn off, less current is passed and the potential between the emitter and the collector rises. This in turn starts to turn Tr2 on. This transistor, when light was falling on the LDR, thus holding Tr1 on, was off since the potential between base and emitter was held at a very low level due to the small potential across Tr1. As soon as Tr2 begins to conduct this increases the current through R4, thus causing Tr1 to cut off even more and in turn switching Tr2 more on. This action takes place very quickly and Tr2 is switched fully on, applying the supply voltage across the bulb causing it to light up.

This switching circuit is known as the Schmitt Trigger and a very useful circuit, it is too, with applications in many fields of electronics. As can be seen, only a few components are required and the switching action, using modern transistors, is very rapid and very positive. As soon as the potential at the base of Tr1 either reaches or falls to a preset level, Tr2 will switch.

The supply necessary to operate the circuit is taken directly from the car battery, though it is good practice to take this via the car's fuse as any

-continued overleaf

No. 30 Automatic light

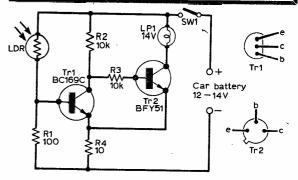


Fig. 1: The circuit of the automatic light with the lead connections of the transistors shown on the right.

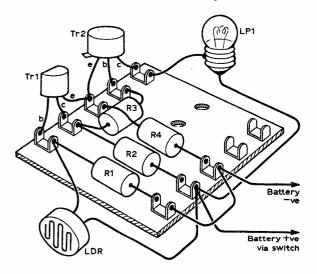


Fig. 2: A suggested component layout on a small piece of tag board.

* components list

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mistakes or serious component failure could create a fire risk.

Car batteries are capable of giving very high currents and therefore the quiescent current is of no great importance but it is in fact low, in the order of 2mA. The recommended bulb draws 40mA and it can be seen that these currents can be supplied by a number of dry batteries, though operation in this manner will not be cheap.

The bulb should last a considerable time as it is being slightly under-run; there is of course a small volts drop across R4 and Tr2.

CONSTRUCTION

A suggested layout is shown in Fig. 2 but this is not critical. One point must be watched however; if the light from the bulb is allowed to fall onto the LDR it will switch the circuit in the same way that sunlight will. This will cause the circuit to oscillate; this can easily be seen as the bulb will flicker or flash. To avoid this the LDR must be pointed away from the bulb.

WAVEMETER MONITOR-by R. F. GRAHAM

continued from page 530

ted. Tuning resonance is shown by maximum meter reading. In driver and similar circuits with small power, the wavemeter needs to be reasonably close to the circuit inductor. For P.A. and high power circuits, it can be at some small distance.

To check the frequency of low level multipliers and similar circuits, or those which are inaccessible, a link line may be used. This is twin flex or co-axial cable about 2ft. long with a loop of two or three turns each end. One loop is held near the equipment inductor, and the other near the wavemeter inductor. Clipping a lead to jack and aerial socket also allows the coil primaries to be used for coupling. Resonance will be shown by a dip in grid current in a subsequent stage.

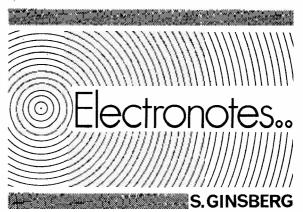
To monitor the audio of an outgoing transmission, plug in phones, tune to resonance and situate the unit where pick-up gives adequate volume.

To tune an aerial, transmitter, tuner or coupling device for maximum radiation from the aerial, a vertical rod or wire is plugged into the aerial socket of the unit, which should be at some distance. Adjustments are then for maximum meter reading, which corresponds to maximum field strength. It is also possible to link couple a dipole or aerial-earth system to the wavemeter for this purpose, or take a short aerial to VCl, ignoring calibration.

The wavemeter produces no signal. It can be used to find bands on a t.r.f. receiver by adjusting the latter until oscillation *just* arises, and holding the wavemeter coil in line with the receiver coil. Resonance when tuning the wavemeter is then shown by the receiver going out of oscillation.

BLUEPRINT SERVICE

We would like to draw readers' attention to the fact that the BLUEPRINT SERVICE has been discontinued and therefore no further BLUEPRINTS are available.



N the beginning there were photocells enclosed in glass envelopes. Variations of these were more sensitive and multi-electrode secondary emission cells appeared like the 931A photomultiplier tube.

In more recent years, many experimenters, and indeed, most of the electronics industry too, has opted for solid state when thinking in terms of a light sensing device. An example here is the ORP12 light-dependent resistor, not to mention the earlier plastic-encapsulated transistors with some of their paint removed!

In America, work is currently proceeding on a technique known as charge coupling for imaging devices. One oxidised surface of a tiny silicon chip is covered with some 288 minute electrodes. These are arranged into groups of three, the last electrode in each group being a common conductor for the whole array.

The area immediately beneath each group of three elements forms a tiny light sensitive device in its own right. Minority carriers are generated in the silicon beneath the electrodes whenever light is focussed onto them. The greater the intensity of the light, the greater the charge, thus forming a direct relationship between light intensity and electrical charge.

The centre electrode of each group is purposely made more positive than the other two elements and thus the charges on minority carriers are attracted to, and congregate at, these centre electrodes.

Individual charges are next moved along the array from one electrode group to the next by putting a successively greater positive voltage on the electrode next to the one which is holding the charge. Eventually all the charges will have been passed along to the end of the device where they are collected by an output electrode.

Because the size of the charge will be dependent upon the amount of original light applied, the final electrode has, in effect, an analogue electrical signal which has a linear relationship to original light patterns of intensity.

How big is a $100,000\mu F$ capacitor—think of the size of some electrolytics. Now imagine how big a one Farad capacitor would be. Just to stretch the imagination even further, think of a 50 Farad capacitor. Fantastic—but they exist and are for sale. Twist in the tale is that the 50F is contained within one-third of a cubic inch. The devices will also retain their storage (if charged with a voltage) and leakage is so low that even after 16 months they will have retained over 97 per cent of the original charge. Think; they could be charged up and used as a form of battery. When discharged, they could simply be recharged again ad infinitum.

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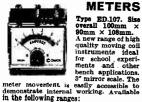
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Current 0-60uA/0 — 12/0
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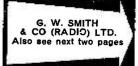


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2N404 20p 2N3605 27p 3N83 2N696 15p 2N3606 27p 3N128	40p BC137 15p BFX87 25p NKT245 20p 70p BC138 37p BFX88 20p NKT261 20p	110p FJH251 25p SN7460 23p CA3019 84p FJJ101 50p SN7472 85p	384 40p 30L15 85p EZ41 45p 384 35p 30L17 80p EZ80 25p
2N697 15p 2N3607 22p 3N140 2N698 25p 2N3638 18p 3N141	77p BC140 35p BFX89 82p NKT262 30p 72p BC141 35p BFX93A 70p NKT264 30p	CA3020 126p FJJ111 50p SN7473 43p FJJ121 60p SN7474 43p	3V4 45p 30P12 80p EZ81 28p 5R4 60p 30P19 80p GZ32 48p
2N699 42p 2N3638A 20p 3N142 2N706 9p 2N3641 18p 3N143	55p BC147 15p BFY11 42p NKT271 20p 67p BC148 11p BFY18 25p NKT262 20p 87p BC149 15p BFY19 25p NKT274 20p	160p FJJ131 60p SN7475 47p CA3021 156p FJJ141 125p SN7476 45p CA3022 180p FJJ181 75p SN7483 87p	5U4 88p 30PL1 70p GZ34 60p 5V4 42p 30PL13 98p KT66 48-05
2N706A 11p 2N3642 18p 3N152 2N708 15p 2N3643 20p 40050 2N709 45p 2N3644 25p 40250	87p BC149 15p BFY19 25p NKT274 20p 55p BC152 17p BFY21 42p NKT275 20p 50p BC153 35p BFY24 45p NKT278 25p	CA3022 180p FJJ181 75p 8N7483 87p CA3023 126p FJJ191 65p 8N7486 50p CA3026 100p FJJ211 125p 8N7490 100p	5Y3 32p 30PL14 90p KT88 22·00 5Z4G 40p 35L6 50p MU14 60p 8/30L2 75p 35W4 80p PABC80 40p
2N709 45p 2N3644 25p 40250 2N718 25p 2N3645 25p 40251 2N718A 30p 2N3691 15p 40309	32p BC154 35p BFY29 40p NKT281 27p 32p BC157 15p BFY30 40p NKT401 37p	CA3028A 74p FJJ251 125p 8N7492 87p CA3028B FJL101 125p 8N7493 87p	8/30L2 75p 35W4 80p PABC80 40p 6AC7 25p 35Z4 80p PC86 60p 6AG7 40p 35Z5 40p PC88 60p
2N726 25p 2N3692 18p 40310 2N727 25p 2N3693 15p 40311	45p BC158 15p BFY41 50p NKT402 90p 85p BC159 15p BFY43 62p MKT403 75p	105p FJY101 25p 8N7495 87p CA3029 87p IC10 250p 8N7496 87p	6AK5 80p 50B5 45p PC97 45p
2N914 17p 2N3694 18p 40312 2N916 17p 2N3702 10p 40314	47p BC160 85p BFY50 22p NKT404 62p 87p BC167 15p BFY51 20m NKT405 75p	CA3029A IC12 250p SN74107 43p 1 165p L900 40p SN74153	6AK6 57p 50C5 40p PC900 43p 6AL5 20p 80 50p PCC84 40p 6AM6 33p 85A2 45p PCC85 40p
2N918 2N9703 10p 40315 2N929 22p 2N3704 12p 40316	87p BC168B 14p BFY52 22p NKT406 62p 47p BC168C 15p BFY53 17p NKT451 62p	CA3080 137p L914 40p 140p CA3035 122p L923 40p SN74154	6AQ5 35p 807 50p PCC88 55p 6AS6 37p 1625 50p PCC89 50p
2N930 24p 2N3705 10p 40317 2N937 52p 2N3706 10p 40319	87p BC169B 14p BFY56A 57p NKT452 62p 55p BC169C 15p BFY76 42p NKT453 47p	CA3036 72p MC724P 66p 220p CA3039 82p MC780P 247p SN74160	6AU6 250 6146 21.50 PCF80 800
2N1090 22p 2N3707 12p 40320 2N1091 22p 2N3708 8p 40323	47p BC170 12p BFY77 57p NKT713 20p 82p BC171 15p BFY90 65p NKT717 42p	CA3041 109p MC788P 82p 180p CA3042 109p MC790P 124p SN74161	6BA6 25p CY31 85p PCF84 50p
2N1131 25p 2N3709 10p 40324 2N1132 25p 2N3710 10p 40326	47p BC172 15p B8X19 17p NKT734 27p BC175 22p B8X20 17p NKT736 35p 80p BC177 22p B8X21 20p NKT773 25p	CA3043 187p MC792P 66p 180p CA3044 120p MC799P 66p SN74164 CA3045 122p MC1303L 220p	6BE6 800 DAF91 250 PCF86 60p 6BH6 450 DAF96 420 PCF800 80p
2N1302 17p 2N3711 10p 40329 2N1303 17p 2N3713 187p 40344 2N1304 22p 2N3714 200p 40347	27p BC178 20p BSX26 45p NKT781 80p	CA3045 122p MC1303L 220p CA3046 81p 262p SN74165 CA3047 137p MC1304P 225p	6BJ6 45p DF91 25p PCF801 50p 6BQ7A 40p DF96 42p PCF802 50p 6BR7 85p DK91 85p PCF805 75p
2N1305 22p 2N3715 222p 40348	52p BC182 12p BSX28 82p OC19 87p	CA3048 204p 275p SN74192 CA3049 180p MC1305P 225p	6BR8 65p DK92 50p PCF806 70p
2N1306 24p 2N3716 290p 40360 2N1307 24p 2N3773 240p 40361 2N1308 25p 2N3791 275p 40362	42p BC182L 10p BSX60 82p OC20 75p 47p BC183 10p BSX61 62p OC22 50p 55p BC183L 10p BSX76 15p OC23 60p	CA3050 185p 386p SN74193 CA3051 134p MC838P 225p	6BW6 85p DK96 42p PCF808 75p 6BW7 70p DL92 85p PCL82 35p 6BZ6 85p DL94 45p PCL83 65p
2N1309 24p 2N3819 84p 40370 2N1507 17p 2N3820 57p 40406	82p BC184 15p BSX77 20p OC24 60p 57p BC184L 12p BSX78 25p OC25 37p	CA3052 165p 549p TAA241 CA3053 46p MC1435P 162p	6C4 83p DL96 42p PCL84 45p 6CD6 \$1:15 DM70 33p PCL85 40p
2N1613 21p 2N3823 75p 40407 2N1631 85p 2N3854 27p 40408	40p BC186 25p BSY24 15p OC26 25p 52p BC187 27p BSY25 15p OC28 60p	CA3054 109p 345p TAA242 CA3055 240p MC1552G 425p	6CL6 56p DY86 33p PCL86 45p 6CW4 63p DY87 35p PFL200 70p
2N1632 30p 2N3854A 27p 40409 2N1637 30p 2N3855 27p 40410	55p BC212L 12p BSY26 17p OC29 60p 62p BC213L 12p BSY27 17p OC35 50p	CA3059 165p 461p TAA243 150p CA3064 120p MC1709CG TAA263 75p	6F1 62p E88CC 65p PL36 55p 6F6G 30p E180F 95p PL81 50p
2N1638 27p 2N3855A 30p 40412 2N1639 27p 2N3856 30p 40467A	50p BC214L 15p BSY28 17p CC36 60p 57p BCY10 27p BSY29 17p CC41 22p	FCH101 85p MFC4000P TAA293 97p TAA300 175p	6F13 88p EABC80 85p PL82 45p 6F14 65p EAF42 85p PL83 45p
2N1701 110p 2N3856A 35p 40468A 2N1711 24p 2N3858 25p 40528	35p BCY30 24p BSY32 25p OC42 25p 72p BCY31 80p BSY36 25p OC44 17p	FCH121 105p 112p TAA310 125p FCH131 50p PA222 487p TAA320 72p	6F15 65p EB91 20p PL84 40p 6F18 45p EBC41 55p PL500 75p
2N1889 32p 2N3858A 30p 40600 2N1893 37p 2N3859 27p 40603	57p BCY32 50p BSY37 25p OC45 12p 50p BCY33 20p BSY38 20p OC46 15p	FCH141 105p PA230 100p TAA350 175p FCH151 105p PA234 100p TAA435 147p	6F23 80p EBC81 80p PL004 80p 6H6 20n EBF80 40n PY32 550
2N2147 72p 2N3859A 32p AC107 2N2160 57p 2N3860 30p AC126	30p BCY34 25p BSY39 22p OC70 15p 20p BCY38 30p BSY43 50p OC71 12p 24p BCY39 80p BSY51 32p OC72 12p	FCH161 50p PA237 185p TAA521 132p FCH171 105p PA246 245p TAA522 360p FCH181 105p PA424 235p TAA530 495p	6J4 50p EBF83 40p PY33 68p 6J5 20p EBF89 82p PY80 85p 6J5GT 80p EBL21 60p PY81 80p
2N2193 40p 2N3866 150p AC127 2N2198A 42p 2N3877 40p AC128 2N2194 27p 2N3877A 40p AC151	24p BCY39 80p BSY51 82p OC72 12p 20p BCY40 50p BSY52 82p OC73 80p 18p BCY41 15p BSY53 87p OC74 80p	FCH191 105p PA264 447p TAA811 445p FCH201 180p PA265 497p TAB101 97p	6J5GT 80p EBL21 60p PY81 80p 6J6 20p EC86 60p PY82 30p 6J7 45p EC88 60p PY83 88p
2N2194A 30p 2N3900 37p AC152 2N2217 37p 2N3900A 40p AC154	22p BCY42 15p BSY54 40p OC75 22p 22p BCY43 15p BSY56 90p OC76 22p	FCH211 180p SN7400 23p TAD100 150p FCH221 180p SN7401 28p TAD110 197p	6J6 20p EC86 60p PY82 30p 6J7 45p EC88 60p PY83 38p 6K8G 35p ECC40 60p PY83 40p 6L6GT 45p ECC84 30p PY800 50p
2N2218 20p 2N3901 97p AC176 2N2219 20p 2N3903 25p AC187	22p BCY54 32p BSY79 45p OC77 30p 25p BCY58 22p BSY90 57p OC78 20p	FCH231 150p SN7402 28p SL403A 187p FCJ101 160p SN7403 28p SL702C 147p	6LD20 40p ECC85 60p PY801 50p 6Q7 40p ECC88 40p U25 75p
2N2220 25p 2N3904 25p AC188 2N2221 25p 2N3905 30p ACY17	27p BCY59 22p BSY95A 12p OC81 20p 27p BCY60 97p C424 15p OC81D 20p	FCJ111 1500 SN7404 280 UA702A 2800 FCJ121 2750 SN7405 280 UA702C 770	68A7 40m ECF80 85m 1726 75m
2N2222 20p 2N3906 30p ACY18 2N2222A 25p 2N4058 15p ACY19	24p BCY70 15p C450 15p OC82 25p 24p BCY71 20p GET102 30p OC82D 15p	FCJ131 275p SN7406 80p UA703C 187p FCJ141 525p SN7408 23p UA709C 125p	68J7 40p ECF86 65p U52 83p 68K7 85n ECH21 57n U191 75n
2N2297 30 p 2N4059 10 p ACY20 2N2368 15 p 2N4060 12 p ACY21	20p BCY72 15p GET113 20p OC83 25p 20p BCY78 20p GET114 15p OC84 25p	FCJ201 100p SN7409 28p UA710C 125p FCJ211 275p SN7410 28p UA716 187p	68L7 85p ECH35 80p U281 40p 68N7 85p ECH42 70p U282 40p
2N2369 17p 2N4061 12p ACY22 2N2369A 17p 2N4062 12p ACY28 2N2410 42p 2N4244 47p ACY39	10p BCY79 30p GET118 20p OC139 25p 17p BCZ10 27p GET120 25p OC140 32p 47p BCZ11 40p GET873 12p OC170 25p	FCK101 480p 8N7411 22p UA723C 162p FCL101 280p 8N7413 35p UA730C 160p FCY101 102p 8N7420 23p UA741C 87p	68Q7 40p ECH81 30p U301 40p 6U4 60p ECH83 40p U801 \$1.00
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2N2539 22p 2N4250 18p ACY44 2N2540 22p 2N4254 42p AD140	47p BD123 80p GET890 22p OC201 50p 47p BD123 80p GET890 22p OC202 75p	PLASTIC 200 PIV 4A 75p	6X5G 80p ECL86 40p UBC81 40p 6X5GT 27p EF37A 60p UBF80 40p
2N2613 27p 2N4255 42p AD149 2N2614 80p 2M4284 17p AD150	47p BD124 75p GET896 22p OC203 40p 62p BD131 75p GET897 22p OC204 40p	ENCAPSULATED 400 PIV 4A 80p 600 PIV 1A 50p 50 PIV 6A 62p 50 PIV 2A 55p 100 PIV 6A 75p	10C2 50p EF39 40p UBF89 35p 10F1 90p EF40 50p UCC84 49p
2N2646 47p 2N4285 17p AD161 2N2711 25p 2N4286 17p AD162	35p BD132 S5p GET898 22p OC205 75p 35p BDY10 125p MAT100 25p OC206 90p	100 PIV 2A 66p 200 PIV 6A 85p 200 PIV 2A 67p 400 PIV 6A \$1.10	10P13 55p EF41 65p UCC85 40p 10P14 41-10 EF42 70p UCF80 55p
2N2712 25p 2N4287 17p AF109 2N2713 27p 2N4288 15p AF114	25p BDY61 125p MAT101 80p OC207 75p 25p BDY61 125p MAT120 25p OCP71 42p	400 PIV 2A 80p	12AT6 30p EF80 25p UCH21 60p 12AT7 30p EF85 35p UCH42 70p
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71		20	PCL82	.32		1.98	U282	.40	AF115	.15	GD9	.20	OC71	.11
	FW4/500.	75	PCL83	.58	R10	.75	U301	.40	AF117	.19	GET113	.20	OC72	.11
	FW4/800.		PCL84	.34	R11	.98	U403	.83	AF121	.30	GET116	.40	OC74	.23
		35	PCL805/8	35		L.75	U404	.38	AF124	.25	GET118	. 20	OC75	.11
		41		.41		.88	U801	.93	AF124 AF126	.20	GET119	.20	OC76	.15
		70		.39	R17					.18			0C77	.27
		48		.65	R19	.30	U4020	.38	AF139	.65	GET573		OC78	.15
				.44	SP42	.75	VP23	.40	AF178	.68	GET587	.43	OC78D	.15
5 6		67	PEN4DD		SP61	.88	VP41	.88	AF180	.48	GET873	.15	0C81	.11
3	HABC80.			.38	TH4B	. 50	VR105	.33	AF186	.55	GET887	.23	OC81D	.11
	HL23DD.		PEN36C		TH233	. 98	VT61A	. 35	AF239	.38	GET897	.23	0C82	iii
	HL41DD.			.35	TP2620	.98	VU111	. 44	BA102	. 45	GET898	.23		.11
	HL42DD.				UABC80		VU120	.60	BA115	.14	M1	.15	OC82D	. 44
	HN309 1.		PEN45D		UAF42	. 49	VU120A		BA116	. 25	M3	.15	OC83	.20
0		53		.75	UBC41	.45	VU133	.35	BA129	.18	MAT100		0C84	.24
		53		.20	UBC81	.40	W76	.34	BA130	.10		. 39	OC123	.23
3	KT2	25	PEN4531	OD	UBF80	. 29	W107	.50	BA153	.15	MAT101		OC139	.23
		75		.98	UBF89	.30	W729	. 60	BC107	.13		.43	OC140	.95
5	KT44 1.		PENDD-		UBL21	. 55	X41	. 50	BC108	.13	MAT120		OC169	.23
2		80		. 88	UC92	. 35	Transisto	76	BC113	. 25		. 39	OC172	.35
8	KT81 2.		PFL200	.53	UCC84	.34			BC115	.15	MAT121		OC200	.22
		63		.47	UCC85	. 34	& Diodes		BC116	. 25		.43	OC201	.38
		63		.44	UCF80	.34	28323	.50	BC118	. 23	OA5	.28	OC202	.43
3	KTW63 .	50	PL81A	.50	UCH21	.60	AA119	. 15	BCY10	.45	OA9	. 13	OC203	.30

.43 OC201 .28 OC202 .13 OC203 .43 OC204

.10 | OC205 .15 | ORP12

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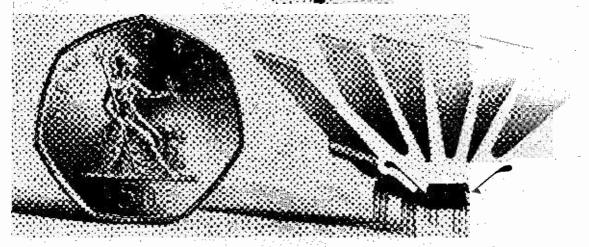
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EBF80	12-2p	PCL83	12√p	6BW7	10p
EBF89	12-∮p	PCF82	12 1 p	6U4	10p
ECC81	10p	PL36	20p	6F23	20p
ECC82	12½p	PL81	17∳p	20P1	20p
ECC83	12-2 p	PY81	7 <u>₹</u> p	20P3	10p
ECL80	7 1 p	PY33	17 ½ p	20D1	10p
EF80	7 ½ p	PY82	7∳p	30P4	20p
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per IC-12



High fidelity Monolithic Integrated Circuit Amplifier

Two years ago Sinclair Radionics announced the World's first monolithic integrated circuit Hi-Fi amplifier, the IC.10. Now we are delighted to be able to introduce its successor, the Super IC.12. This 22 transistor unit has all the virtues of the original IC.10 plus the following advantages:

- 1. Higher power,
- 2. Fewer external components.
- 3. Lower quiescent consumption.
- 4. Compatible with Project 60 modules.
- 5. Specially designed built-in heat sink. No other heat sink needed.
- 6. Full output into 3, 4, 5 or 8 ohms.
- 7. Works on any voltage from 6 to 28 volts without adjustment.
- 8. NEW 22 transistor circuit,

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Output power 6 watts RMS continuous (12 watts peak).

Frequency Response 5 Hz to 100KHz ± 1 dB

Total Harmonic Distortion Less than 1%. (Typical 0.1%) at all output powers and all frequencies in the audio band.

Load Impedance 3 to 15 ohms.

Power Gain 90dB (1,000,000,000 times) after feedback.

Supply Voltage 6 to 28 volts (Sinclair PZ-5 or PZ-6 power supplies ideal).

Size 22 x 45 x 28 mm including pins and

Input Impedance 250 Kohms nominal.

Quiescent current 8mA at 28 volts.

With the addition of only a very few external resistors and capacitors the Super IC.12 makes a complete high fidelity audio amplifier suitable for use with pick-up, F.M. tuner etc. Alternatively, for more elaborate systems, modules in the Project-60 range such as the Stereo 60 and A.F.U. may be added. The comprehensive manual supplied with each unit gives full circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include car radios, oscillators etc. The very low quiescent consumption makes the Super IC.12 ideal for battery operation.



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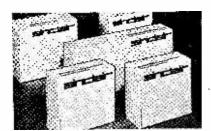
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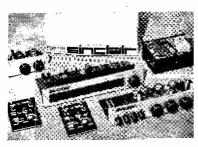
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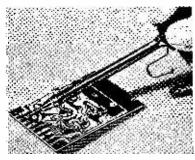


Sinclair Project 60

The World's leading range of high fidelity modules

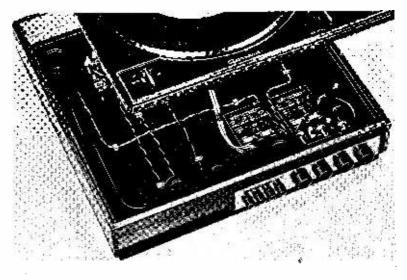






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Project 60 offers more advantage to the constructor and user of high fidelity equipment than any other system in the world.

Performance characteristics are so good they hold their own with any other available system irrespective of price or size.

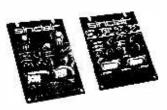
Project 60 modules are more versatile — using them you can have anything from a simple record player or car radio amplifier to a sophisticated and powerful stereo tuner-amplifier. Either power amplifier can be used in a wide variety of applications as well as high fidelity. The Stereo 60 pre-amplifier control unit may also be used with any other power amplifier system, as can the AFU filter unit. The stereo FM tuner operates on the unique phase lock loop principle to provide the best ever standards of sensitivity and audio quality. Project 60 modules are very easily connected together by following the 48 page manual supplied free with all Project 60 equipment. The modules are great space savers too and are sold individually boxed in distinctive white and black cartons. With all these wonderful advantages, there remains the most attractive of all — price. When you choose Project 60 you know you are going to get the best high fidelity in the world, yet thanks to Sinclair's vast manufacturing resources (the largest in Europe) prices are fantastically low and everything you buy is covered by the famous Sinclair guarantee of reliability and satisfaction.

Typical Project 60 applications

System	The Units to use	together with	Cost of Units
Simple battery record player	Z.30	Crystal P.U., 12V battery volume control	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U. volume control etc.	£9.45
20 + 20 W. stereo amplifier for most needs	2 x Z.30s, Stereo 60, PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
20 + 20 W. stereo amplifier with high. performance spkrs.	2 x Z.30s, Stereo 60, PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc.	£26.90
40 +40 W. R.M.S. de-luxe stereo amplifier	2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43

from a simple amplifier to a complete stereo tuner amplifier with Project 60 modules

Z.30 & Z.50 power amplifiers



The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at full output and all lower outputs. Whether you use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.

SPECIFICATIONS (2.50 units are interchangeable with Z.30s in all applications). **Power Outputs**

Z.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts. **Z.50** 40 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts

Frequency response: 30 to 300,000Hz±1dB. Distortion: 0.02% into 8 ohms.

Signal to noise ratio: better than 70dB unweighted. Input sensitivity: 250mV into 100 Kohms. For speakers from 3 to 15 ohms impedance.

Size: 14 x 80 x 57 mm

Built, tested and guaranteed With circuits and instruc-

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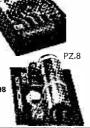
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Power Supply Units

Designed special for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 where a stabilised supply is essential.

PZ.530 volts unstabilised £4.98 PZ.635 volts stabilised £7.98 PZ.8 45 volts stabilised

ss mains transformer) £7.98 PZ.8 mains transformer £5.98



The Sinclair Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.

Project 60 Stereo F.M. Tuner





Firet in the world to use the phase lock loop principle

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. Good reception is possible in difficult areas, and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high fidelity system

SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz. Capture ratio: 1.5dB. Sensitivity: 2nV for 30dB quieting: 7nV for full limiting. Squelch level: 20pV. A.F.C. range: ±200 KHz. Signal to noise ratio: > 65dB. Audio frequency response: 10 Hz – 15 KHz (±1dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: 2nV. Cross talk: 40dB. Output voltage: 2 x 150mV R.M.S.

Operating voltage: 25-30 VDC. Indicators: Mains on; Stereo on; tuning. Size: 93 x 40 x 207 mm.

Built and tested. Post free.

£25

Stereo 60 Pre-amp/control unit



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

SPECIFICATIONS-Input sensitivities: Radio - up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A SPECIFICATIONS—Input sensitivities: Hadio—up to 3mV, wag, p.u. sinv: correct to n.i.A.A curve ±148 i.20 to 25,000 Hz. Ceramic p.u. - up to 3mV Aux—up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE + 15 to —15dB at 10 KHz: BASS ± 15 to —15dB at 100Hz. Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm. £9.98 Built tested and guaranteed.

Built tested and guaranteed.

A.F.U. High & Low Pass Filter Unit



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less

loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages – rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V. Current - 3mA. H.F. cut-off (-3dB) variable from 28KHz to 5KHz. L.F. cut-off (-3dB) variable from 25Hz to 100Hz. Distortion at 1 KHz (35V. supply (0.02% at rated output. Size: 66 x 40 x 90 mm.

Built tested and guaranteed.

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ance	Volt- in	ina-		Post	ance	Volt-			Price	Post
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in μF										
. 4	370V 24 ×	$2 \times 1T/E$	0.40	0-08	4,000	50V	41×11			0.08
7	370∇ 21×	$9 \times 2T/E$	0.40	0.09	5,000	55 V	44 X 14	8/T	0.60	0.12
10	600V 4×3			0.10	6,600	50V	41×21	SIT	0.65	0.12
32		K1 T/E		0.05	7,000	50V	$4\bar{1} \times 2\bar{1}$	S/T	0.75	0.12
32		K14 C	0.80	0.05	8,000	50V	41×21	S/T	0.75	0.12
50		1 W/E	0.20	0.05	9,200	35V	41×21	S/T	0.50	0.12
60		kî≇ C	9-80	0.05	10,000	70V	41×21	C		0.12
80		<1 W/E	0.80	0.05	13,500	38₹	41 × 21	S/T	0.60	0.12
100		<1 € C	0.85	0.05	14,000	35V	41×3	S/T	0.60	0.15
1.000	70V 3	xîž C	0.35	0.05	20,000	55 V	6×3	S/T	1.25	0.18
1,200		×21 C	2.00	0.15	33,000	25 V	41×21			0.12
1,500	50∇ 4½ ×	K1∯ C	0.35	0.08	34,800	40V	6×3			0.18
2,000		x1 C	0.80	0.05	100+200	300V	4½×1%			0.08
2,500	50V 4	x1ŧ C	0.40	0.07	100+400	275V	4×1#	P.C.		0.08
3,000		×24 C	0.60	0.15	200+200	350∇	41×11	C	0.50	0.12
3,600	20V 41	x1 8/T	0.50	0.12						

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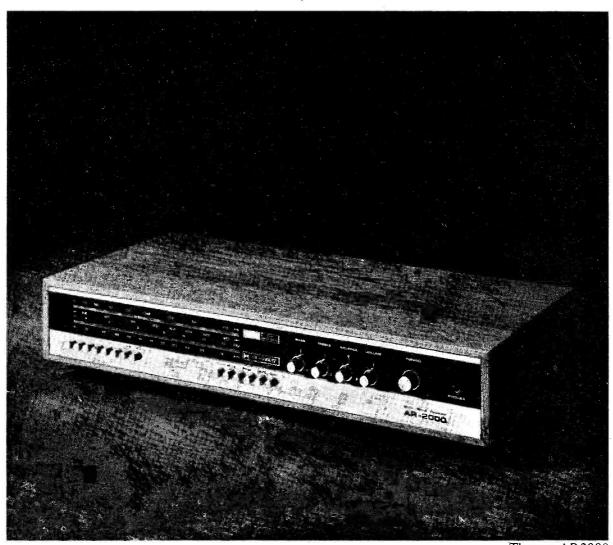
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LEAK Stereo 70 Chassis	69.00	52.00
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PIONEER SA900	134 - 10	95.95
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ROGERS Ravensbourne	59 - 50	45·95 49·50
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8 electronically mixed inputs		139 - 00
All the above take both ceramic	and m	agnetic
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TUNERS		
TURERS	E2.78	49.00

cartridges.		
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	86-50	52.00
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All above Tuners are complete with		
Decoder except where starr		Stelen
Decode, except where stati-	eu.	

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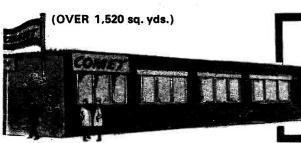
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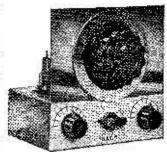
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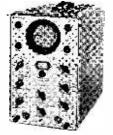
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RT38 0.80 PT.302 0.85 0.80 0.90 1.10 0.80 0.95 0.60 0.83 0.40 0.83 0.85 0.85 U801 UABC80 OC3 0.38 EK90 EL33 EL34 EL36 0.30 1.25 0.50 0.50 PL502 PL504 PL508 PL509 6F23 6F24 6F25 OD3 1B3GT 10LD11 0.65 0.40 0.50 0.55 0.55 0.40 0.40 7.00 0.20 10P13 10P14 0.80 1.10 0.70 0.50 0.55 0.55 0.50 0.50 0.80 UAF41 UAF42 1CP31 PL509 PL801 PL802 PM84 PY31 PY33 PY80 PY81 10P14 12AB5 12AC6 12AD6 12AE6 12AL5 12AQ5 12AT6 12AT7 1L4 1R4 1R5 184 185 1T4 1U4 BRAND 6F26 UBC41 UBC81 UBF80 UBF89 UBLI | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | 0.00 | EL37 0.45 EL41 EL41 EL42 EL81 EL83 EL84 EL85 EL86 UBC81 0.40 UBF80 0.40 UBF80 0.85 UBL1 0.60 UBL21 0.65 UC92 0.40 UCC85 0.40 UCF80 0.55 UCH21 0.60 UCH42 0.70 UCH81 0.40 UCL81 0.80 UCL82 0.35 UCL82 0.36 **ELECTRONIC VALVES** 0.80 0.60 0.50 0.50 0.38 0.40 PY81 0.30 PY82 0.35 PY83 0.38 PY88 0.40 PY500 1.00 PY800 0.40 PY801 0.50 PZ30 0.35 QQVO2-6 12AT7 0.85 12AU6 0.85 12AU7 0.80 12AV7 0.65 12AX7 0.75 12AX7 0.75 12BA4 0.60 12BA6 0.40 12BA7 0.40 12BB7 0.40 12BB7 0.60 12BY 0.60 12BY 0.60 12BY 0.60 12BY 0.60 30C17 0.90 30C18 0.80 30F5 0.85 30FL1 0.75 90AG 90AV 90CI 90CV 2.40 2.50 0.65 2.40 1U5 1V2 0.83 0.85 0.65 1.00 ECC189 0.85 ECP80 0.35 ECP80 0.35 ECP80 0.85 ECP8041.85 ECH81 0.30 ECH81 0.30 ECH82 0.45 ECH81 0.50 ECH82 0.35 ECL83 0.45 ECL83 0.45 ECL83 0.50 ECL83 0.50 ECL83 0.50 ECL83 0.50 ECL85 0.50 ECL85 0.50 ECL85 0.50 ECL85 0.50 ECL85 0.50 ECL85 0.50 EF42 0.50 EF42 0.70 EF83 0.55 DY802 E88CC TET.90 1X2B 2D21 EL95 EL821 EL822 ELL80 EM34 EM71 EM80 EM81 EM85 EM85 EM87 E180F E180F 1.00 E810F 2.90 EABC800.35 EAF42 0.55 EBC33 0.50 EBC41 0.55 EBC81 0.30 3A4 3BP1 884 3V4 30FL12 1.20 30FL14 0.85 30L1 0.40 30L15 0.85 30L17 0.80 30P12 0.80 30P11 0.75 30PL13 0.93 30PL14 0.90 35A3 0.55 35A5 0.75 35B5 0.65 30FL12 1 20 807 0.50 8.75 0.75 0.70 1.60 1.25 2.25 0.80 2.00 1.25 0.40 0.55 0.55 813 866A 5642 6080 6146 3.00 0.35 QQV02-6 QQV03-10 1.25 QQV08-20-A QQV08-20-A TT21 3.00 TT22 3.20 TY2-125 U18/20 0.75 U25 0.80 U31 0.60 U37 2.10 UCLS2 0.38 UCLS2 0.36 UCLS3 0.60 UF41 0.80 UF42 0.80 UF85 0.38 UF85 0.38 UF85 0.40 UF89 0.40 UF89 0.40 UY1N 0.80 UY1N 0.8 3V4 0.48 5R4GY 0.75 5R4GY 0.75 5U4G 0.35 5V4G 0.45 5Y3GT 0.40 5Z3 0.60 5Z4G 0.40 0.55 0.80 0.40 0.40 0.50 0.60 0.60 0.38 0.60 EBC81 EBF80 EBF83 EC53 EC86 EC88 EC90 EC91 6939 7199 7360 7586 7895 9002 9003 0.55 0.60 0.55 12BY7 0.60 12K5 0.70 12K7GT0.40 12Q7GT0.40 12SR7 0.40 20D1 0.50 20L1 1.10 20P1 0.50 20P4 1.10 5Z4G 0.40 6/30L2 0.80 6CL6 0.55 6CW4 0.86 6CY5 0.45 6CY7 0.70 6D3 0.50 6DC6 0.80 6DK6 0.50 6DQ6B 0.70 6DS4 0.75 6EA8 0.60 6EH7 0.30 6EJ7 0.35 6EW6 0.70 6CL6 6CW4 6AB4 6AF4A 6AG7 6AH6 6AJ8 EY51 EY80 0.85 0.55 0.40 68N7G1 68Q7 68R7 6T8 6U4GT 6U8A 6V6GT EY80 EY81 EY83 EY86 EY87 EY88 EZ40 EZ41 35C5 0.50 35D5 0.75 35L6GT0.50 9003 0.50
AZ1 0.55
AZ31 0.55
CBL1 0.90
CBL31 1.00
CY31 0.85
DAF96 0.45
DF96 0.45
DK40 0.60
DK92 0.55
DL96 0.45
DM160 0.65
DY86 0.32 EC91 EC92 EC93 EC8010 ECC40 ECC81 ECC82 0.50 0.30 0.85 0.55 2.25 0.85 0.85 35L6GT 0.50 35W4 0.85 35Z3 0.70 35Z4G 0.85 35Z5GT 0.60 50A5 0.75 50B5 0.50 50C5 0.50 6AK5 0.35 0.40 0.40 1.10 20P5 25C5 20P5 1.20 25C5 0.50 25L6GT 0.50 25Z4G 0.30 25Z6GT 0.85 30A5 0.50 30AE3 0.40 30C1 0.30 30C15 0.80 6X4 0. 6X5GT 0. 6X5GT 0. 6X8 0. 6Y6G 0. 7Y4 0. 9BW6 0. 0.43 0.20 0.35 0.38 0.38 0.45 0.45 0.40 0.85 0.40 0.60 0.70 0.65 0.60 0.50 6AL3 U37 U52 U76 U78 U191 U201 2.10 0.35 0.35 0.35 0.75 6AL5 6AM5 6AM6 6AQ6 6AQ6 EF85 EF86 EF89 EF91 EF92 EF95 EF97 0.30 0.30 0.30 0.40 0.40 0.50 0.20 0.35 0.30 0.28 0.33 0.35 0.35 EZ80 EZ80 EZ81 GY501 GZ30 GZ31 ECC85 ECC85 ECC89 ECC91 0.70 0.75 0.75 0.35 6F1 6F5 50CD6G1 90 6AR5 10C2 10D1 50L6GT0.55 85A2 0.50

7	R	Α	Ν	S	IS	T	DRS

6F6G

TRANSISTORS						
2N696	0.17	AC127	0.20	BD121	0.65	
2N697	0.17	AC128	0.15	BD123	0.80	
2N698	0.30	AC132	0.35	BD131	0.85	
2N705	0.76	AC153	0.20	BD132	0.85	
2N706	0.10	AC154 AC157	0.15 0.20	BD135	0.80	
2N708 2N753	0.15 0.25	AC169	0.20	BF115 BF167	0.20 0.18	
2N929	0.23	AC176	0.25	BF173	0.20	
2N930	o ok	AC187	0.80	BF178	0.25	
2N987	0.30	AC188	0.80	BF179	0.80	
2N1131	0.25	ACY17	0.275	BF180	0.35	
2N1132	0.25	ACY18	0.20	BF181	0.25	
2N1184	1.25 0.40	ACY19	0.25	BF184	0.25	
2N1301	0.40	ACY20 ACY21	0.20 0.20	BF185	0.20	
2N1302 2N1304	0.25	ACY22	0.20	BF194 BF195	0.15 0.15	
2N1305	0.25	AD140	0.50	BF196	0.20	
2N1306	0.25	AD149	0.50 0.35	BF197	0.20	
2N1307	0.39	AD161	0.35	BF200	$0.20 \\ 0.35$	
2N1308	0.25	AD162	0.35	BFW87	0.25	
2N1309 2N1613	0.30 0.22	ADZ11 ADZ12	1.25 1.25	BFW88 BFW89	0.28	
2N1013	0.25	AF114	0.20	BFW91	0.20	
2N1756	0.75	AFI15	0.20	BFX88	0.25	
2N2147	0.75	AF116	0.20	BFY17	0.40	
2N2160	0.65	AF117	0.20	BFY19	0.60	
2N2217	0.30	AF118	0.45	BFY50	0.25	
2N2218	0.80	AF125	0.25	BFY51	0.20	
2N2219	0.35	AF127	0.20	BFY52	0.25	
2N2369A 2N2477	0.20	AF180 AF181	0.35 0.35	BSY26	0.20 0.20	
2N2411 2N2646	0.60	AF186	0.40	BSY27 BSY28	0.20	
2N2905	0.35	AF239	0.40	BSY65	0.20	
2N2923	0.15	AFZ11	0.45	BSY95A	0.15	
2N2924	0.15	ASY26	0.25	OC16	0.50	
2N2926	0.125	ASY27	0.80	OC22	0.50	
2N3053	0.25	ASY28	0.25	OC23	0.60	
2N3054 2N3055	0.60 0.75	ASY29 ASY54	0.80 0.25	OC24	0.60	
2N3133	0.75	ASZ15	0.70	OC25 OC26	0.85 0.25	
2N3134	0.30	ASZ16	0.70	OC28	0.60	
2N3391	0.20	ASZ17	0.75	OC29	0.60	
2N3392	0.15	ASZ18	0.75 0.25	OC30	0.75	
2N3393	0.15	ASZ20	0.25	OC35	0.50	
2N3394	0.15	ASZ21	0.40	OC36	0.60	
2N3395 2N3402	0.20	BC107 BC108	0.125 0.125	OC42	0.20	
2N3402 2N3403	0.15 0.15	BC109	0.125	OC44 OC45	$0.20 \\ 0.15$	
2N3404	0.35	BC113	0.25	OC70	0.10	
2N3414	0.20	BC118	0.80	OC71	0.12	
2N3415	0.15	BC134	0.30	OC72	0.25	
2N3416	0.25	BC147	0.175	OC73	0.80	
2N3417	0.25	BC148	0.15	OC75	0.20	
2N3702	0.12	BC149 BC152	0.15	OC76	0.20	
2N3703 2N3704	0.12 0.17	BC152 BC158	0.15 0.15	OC78	0.25	
2N3704 2N3707	0.17	BC175	0.20	OC78D OC81	0.20 0.25	
2N3709	0.12	BC186	0.25	QC81D	0.15	
2N3710	0.12	BCY30	0.25	OC83	0.20	
2N3819	0.35	BCY31	0.40	OC139	0.80	
2N3906	0.20	BCY33	0.25	OC140	0.85	
28702	0.50	BCY34	0.80	OC141	0.60	
28746	0.25	BCY72	0.20 0.80	OC170	0.25	
AC113	0.15	BCZ10	0.80	OC171	0.25	

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GZ32 GZ33

Type U4312 U4313

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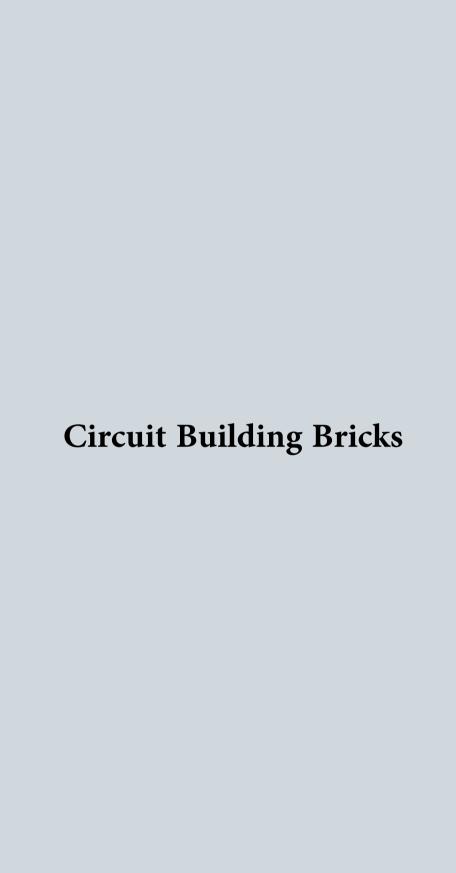
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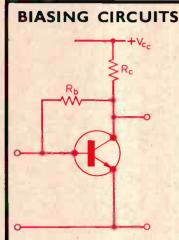


CIRCUIT BUILDING BRICKS

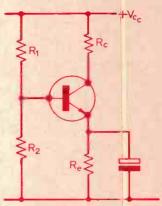
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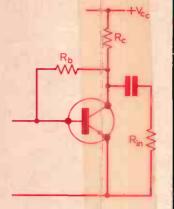
Working out the bias resistors. This circuit is suitable for silicon transistors where l_{cbo} is usually negligible. Collector voltage is about half the supply voltage V_{cc} . So $V_{ce} = V_{cc} - I_c R_b$. If we know the d.c. current gain (hFE), we can work out Rb as h_{FE}R_c, choosing the mid-point of the manufacturer's quoted figures. Collector current is approximately emitter current and base current is the voltage dropped across the load resistor, divided by R_b+h_{FE}R_c.



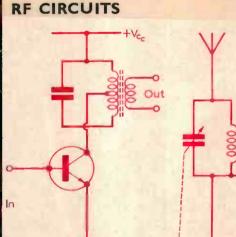
For germanium transistors (with a significant collector/ base leakage current) this arrangement may be used. Remember, base-emitter voltage of silicon transistors at low current is 0.7V, but for germanium transistors, 0.1 to 0.2 volts. Here, we can say:

$$V_{b} = \frac{R_{2}V_{cc}}{R_{1} + R_{2}}$$

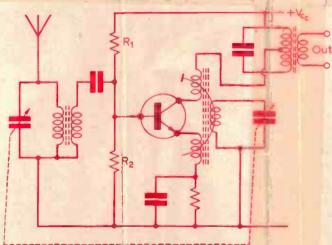
Input impedance is increased by current feedback, via R_e. Voltage across R_e is proportional to collector cur-



Coupling two stages this way makes the input resistance of the second stage effectively the load resistance of the first. The value of coupling capacitor depends on the low frequency response of the amplifier. The base bias resistors are in parallel with the input resistance and with the emitter resistance plus internal emitter resistance of the transistor. Voltage-derived feedback is obtained from Rb, effectively reducing

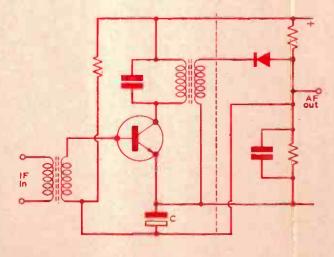


Typical circuit of r.f. amplifier, with the collector tapped down the tuned circuit to preserve matching. Blas will be obtained from a.g.c. applied to the input circuit.



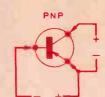
A typical self-oscillating mixer circult. Oscillator feed-back for the emitter is taken through low-impedance coils from the collector. The transistor is biased to Class A by R₁, R₂. As oscillation

rises, rectification of the oscillating signal occurs through the base/emitter diode. Loop gain decreases after each cycle, stabilising the amplitude, causing quiescent emitter current to rise.

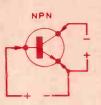


A.G.C. is derived from the detector circuit (to the right of the dotted line) and applied to the input of the first i.f. amplifier. A large electrolytic capacitor, C, is provided to decouple a.f. from the a.g.c.

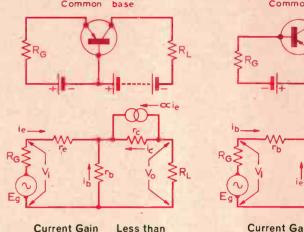
The i.f. transformer may be damped by a diode to improve bandwidth characteristics. The stage shown is for a.m. It is not so necessary to provide a.g.c. for f.m. i.f. stages, where limiting is already applied.



P.N.P. transi shown are rela P-region, po emits holes i region). Colle is negatively b ing holes from



N.P.N. transi shown are rela N-type emitter tively biased v base, and coll positive. N-typ jects free ele type base.



Current Gain

Input Resist-

ance **Output Resist**ance

Voltage Gain

unity Low (200Ω)

High (200kΩ) High (about

Current Gain Input Resistance **Output Resist-**

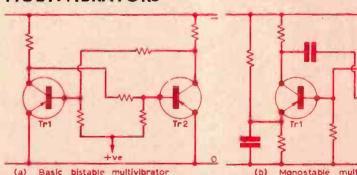
ance

Voltage Gain

COMMON TRANSISTOR SYMBOLS A

V _{be}	=	Base/emitter voltage
Vce	=	Collector/emitter voltage
16	=	Base current
l _c β	=	Collector current
	=	Collector/base current ratio
α	=	Collector/emitter current ratio
α'	2000	$\frac{\alpha}{(1-\alpha)} \left[\alpha = \frac{\alpha'}{(1+\alpha')} \right]$
V _{CBO}	-	Max. collector/base voltage
V _{CBO}	-	Max. collector/emitter voltage
V _{CBO}	-	Max. emitter/base voltage
Icm		Max. peak collector current
Сво	100	Max. collector/base reverse curre
IEBO	=	Max. emitter/base reverse currer

MULTIVIBRATORS



Three types of transistor switched device are in common use as pulse generators. When a transistor remains in a given state until compelled to change by an external 'trigger' signal, it is termed 'stable'. If, when placed in this state, it then reverts, after a while, to Its former state without further external stimulus, it is termed 'unstable'.

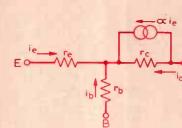
Another class of circuit is free-running; that is, it does not stay in either state but automatically switches be-tween 'bistable' when an external trigger pulse is needed to switch it and 'monostable' when it always reverts to its former state after triggering.

This type is called 'astable'. The astable circuit switches at a frequency determined by the circuit time constants. It can be syn-chronised by an external signal. The multivibrator can be made monostable, bistable or astable by alteration of blas conditions.

in (a) the basic bistable circuit is given. When one transistor is conducting, its collector potential causes the other to be cut off. An external signal causes a change in state. A second external signal causes the circuit to revert to the first state. A recurrent triggering signal will cause a rectstor. Polarities tive to emitter. sitively-biased nto base (nctor, p-region iased, collectbase.

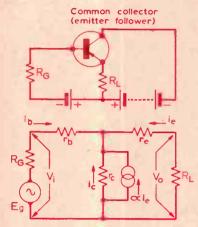
stor. Polarities tive to emitter. must be negaith respect to ector is made e emitter inctrons into p-

mitter



Basic T arrangement of transistor internal resistances

Basic T equivalent circuit



Current Gain High (50) Input Resist-High (100k Ω) ance **Output Resist-**

Low (1k Ω) ance Voltage Gain Unity

ND ABBREVIATIONS

VCEO(sus) VcE(sat) he

Ico

High (50)

Medium $(1k\Omega)$

Medium

 $(40k\Omega)$

High (about 250)

Collector/emitter sustaining voltage Collector/emitter saturation voltage

D.C. current gain

Forward current gain, common emitter, output shorted

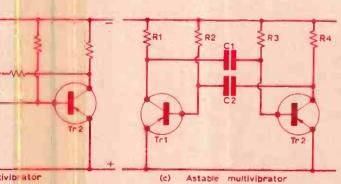
Reverse voltage feedback ratio, common emitter, input open

Input resistance, common emitter, output shorted

Output admittance, common emitter, input open

Collector leakage current (grounded emitter)

Collector leakage current (grounded base)



angular voltage waveform to appear at each collector, at half the frequency of the triggering signal.

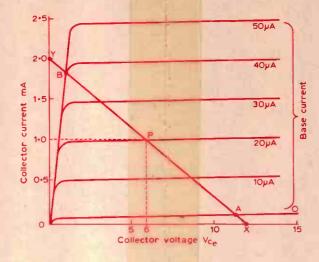
In (b) is shown the monostable multivibrator circuit, similar to the bistable, except for a change of blas and a capacitor coupling. In the stable state, Tr1 is off and Tr2 on; in its unstable state, Tr2 is off and Tr1 on. The duration of the unstable state is the time the base potential of the non-conductive transistor takes to fall to zero.

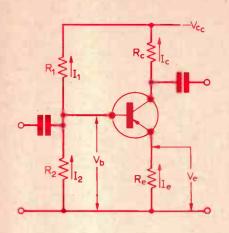
If both the inter-stage couplings are made capacitive, an astable multivibrator is obtained. In (c) there is no

emitter bias, both bases returned via resistors to the supply line. If R=R2=R3, and C=C1=C2, the multivibrator is symmetrical. As a result, a square waveform should be produced, at both collectors, with a periodicity T=1.38RC. The free-running frequency is the reciprocal of T, i.e.

1:38RC An example of this type of circuit used as an oscillator in a 405-line television receiver, free-running frequency 7.5kHz, triggered frequency 10.125kHz, would have R1 and R4, 3kΩ; R2 and R3, $100k\Omega$; C1 and C2 0.001 µF. A number of transistors will perform adequately in this kind of circuit.

DC CONDITIONS





Choosing the right transistor is made easier by reference to its operating conditions. This is a series of curves for different base currents, with collector current plotted against collector voltage. Joining the supply line voltage to the saturation current (quoted in maker's literature) produces a 'load line'. Choose a point midway along this—P in the diagram. Then, in the example, a base current of 20 microamps is required for a collector current of 1mA and a collector voltage of 6 volts. Multiplying the base current for d.c. conditions by the current gain of the transistor gives the required value of collector current.

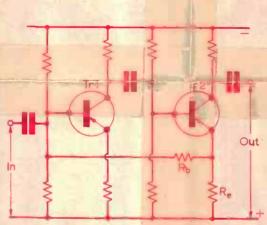
The potentiometer, R₁, R₂, across the supply, provides a tapping point for base control voltage. A bit more bias can be provided by the negative feedback of $R_{\rm e}$, in series with the emitter. Ideally, R₁ and R₂ should be as low as possible, but this

imposes a drain on the the collector current. supply, so a compromise solution is reached.

For a typical stage with a supply voltage of 6V, the values of R1 and R2 would be 23kΩ and 5kΩ respectively (for a transistor with a current gain of about 50) and Re would be around 1kΩ. Choice of values is partly determined by the transistor, partly by the power supply. R₁, R₂ must provide a bleed current about 1mA. ten times that of the base current, but about a fifth of

Silicon transistors, with little collector leakage current and a higher 'starting' voltage, will have higher values of R_e. For a typical stage with a supply of 7 volts (positive, this time, for the n.p.n. transistor), R1 and R2 would both be 15kΩ. Remight be $2.7k\Omega$ for a transistor with an h_{FE} of 100 (50-150) using an emitter current of

AF CIRCUITS



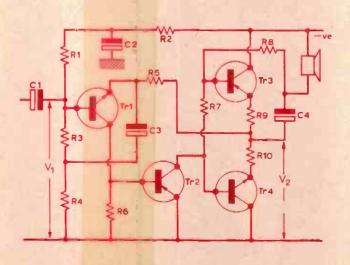
The basis of much simple audio design-a two-stage current amplifler. The curfeedback is rent amplifier gives a current dependent. output for a voltage input and has a low input resistance and a high cutput resistance. D.C. stabilisation is included as before and feedback is obtained by Rb from the emitter of Tr2 to the base of Tr1. Overall current gain can be given by R_b/R_e and is independent of transistor

parameters. The emitter of

Tr1 is decoupled for a.c. negative feedback-i.e., the frequency-

For a voltage amplifier (high input resistance, low output resistance), it is more usual for the feedback Rb to be taken from the collector of Tr2 to the emitter of Tr1. The emitter of Tr2 is decoupled and overall voltage gain is again Rb/Re, but this time R, is the emitter resistor Tr1 by Tr2. A circuit equivalent shows the way this system operates, with the 'collector' current the sum of Tr1 coilector current and Tr2 base current. Bias stability is

Direct coupling of the two-stage amplifier saves com-supply voltage of 20 volts, the supply voltage of 20 volts, the ponents by stabilisation of only practical modification would be a further stage of decoupling between the supply to Tr1 and the main supply, which would be positive. This might consist of a 100kΩ resistor with a In a typical pre-amplifier Re1, 680Ω ; Rc2, $12k\Omega$; Re2, using sillcon n.p.n. transistors, such as the BC109 or



The basic complementary push-pull circuit has the signal from the driver transistor Tr2 simultaneously driving the bases of a pair of transistors, one of which Is p.n.p. and the other, n.p.n. These conduct during alternate half-cycles of signal. Tr1 is a pre-amplifier and a difference-amplifier to compare the voltages of input and output.

The quiescent bias voltage for the output transistors is developed across R7. Bias is set for a linear combined transfer characteristic, to reduce crossover distortion. In practice, compensation for temperature changes would be made by fitting a thermistor across R7, and for supply changes by the use of diodes in this part of the circuit.