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7403 10p 74121 25p CA3060 7404 12p 74122 35p CA3065 7405 12p 74123 40p CA3076	225p SL917B 65)p 200p SN76003N 150p 250p SN76013N 110p	AC126 20p BC AC127 20p BC AC127/01 25p BC	178 15p BF120 50 182 10p BF121 45 182L 12p BF123 45	DP BY127 15p 5p BY133 25p 5p BY164 50p	Receive. £2 00 each,	£3 50 pair. Data 10p
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7408 12p 74120 12op CA3086 7409 12p 74130 12op CA3086 7410 12p 74131 90p CA3088	60p SN76033N 150p 190p SN76227N 160p	AC153 300 BC AC153K 400 BC AC154 300 BC	184L 12p BF154 18 186 20p BF160 18	p E100 42p p E300 47p	4 WAY. 4 POLE 3 W.	AY. All at 40p Each.
7411 15p 74132 45p CA3089 7412 15p 74135 90p CA3090A 7413 25p 74136 80p CA3123E	AQ 360p SN766228N 180p 130p TAA300 100p	AC187 20p BC AC188 20p BC AC188 20p BC	204 12p BF1701 10 205 12p BF173 20 207 12p BF178 2	P E310 60p DP E420 180p EP E430 120p	OPTO ELECT	RONIC
7414 45p 74137 90p CA3130 7416 25p 74138 100p CA3140 7417 25p 74141 50p LF356	100p TAA350 190p 60p TAA550 35p 80p TAA570 220p	ACY19 35p BC ACY20 35p BC ACY20 35p BC	212 11p BF179 23 212L 12p BF180 30 213 12p BF181 30	Dp MJE340 45p MPSA05 30p MPSA06 32p		
7420 12p 74142 180p LF357 7421 20p 74143 270p LM211H 7421 20p 74144 270p LM300T5	80p TAA661B 140p 250p TAA700 350p 170p TAA790 350p	ACY40 50p BC ACY41 50p BC ACY41 50p BC	213L 15p BF182 30 214 15p BF183 29 214L 18p BF184 20	0p MPSA56 32p 5p TIP29 35p	0.125 or 0.2 in 15p each. 10 fo	ch RED LED\$ pr £1 00, 100 for
7422 150 74145 550 LM301A1 7423 200 74147 1000 LM304 7425 200 74148 900 LM307	X 30p TAD100 150p 200p TAD110 130p 65p TBA120S 60p	AD130 150p BC AD143 150p BC	237 10p BF186 25 237B 15p BF194 10 268 16p BF195 10	5p TIP29A 40p 0p TIP29C 40p 0p TIP30 35p	DL747 Seven S Anode Display	egment Common segment Common ys: Character
7426 22p 74150 65p LM308TC 7427 22p 74151 45p LM308TC 7428 25p 74153 45p LM308DI	05 100p TBA120T 85p L 100p TBA480Q 200p TBA520Q 200p	AD149 80p BC AD161 30p BC AD162 30p BC	294 30p BF196 10 300 25p BF197 10 301 25p BF198 29	0p TIP30A 40p 0p TIP30B 40p 5p TIP30C 45p	FND500 Seven Common Cath	t∙00 each. Segment ode Displays.
7430 12p 74154 70p LM310T C 7432 20p 74155 45p LM310T C 7433 28p 74155 45p LM311T C	05 150p TBA530Q 200p 05 150p TBA540 200p 150p TBA550Q 250p	AD161/2MP BC 70p BC AF114 25p BC	303 30p BF199 2 307 15p BF200 30	5p TIP31 40p 0p TIP31A 45p 0p TIP31B 50p	Character Heig £1 30 each. 4 2N5777 Photo-I	aht 0·5″ for £5 00. Darlington
7437 20p 74157 45p LM317A 7438 29p 74160 55p LM324 7440 12p 74161 55p LM339	70p TBA560C 250p 60p TBA641A12	AF115 25p BC AF116 25p BC AF116 25p BC	317 12p BF225 20 318 12p BF241 10	0p TIP31C 55p 0p TIP32 40p 6p TIP33 60p	60p each. 0 125 or 0·2″ ∖ Green LEDS	fellow and 15p. each. 10 for
7441 45p 74162 55p LM348N 7442 40p 74163 55p LM380 7443 60p 74163 55p LM381N	90p TBA700 180p 90p TBA720Q 225p	AF178 95p BC AF178 95p BC AF139 35p BC	323 300 BF244B 3 328 180 BF255 2 337 200 BF257 2	5p TIP34 70p 5p TIP35 200p 5p TIP35 200p	£1 40. 100 for 7730 160p. HP5 OPP12 Japage	£12 00. HP5082- 5082-7750 160p.
7444 60p 74164 60p LM382 7445 65p 74165 60p LM391 7445 65p 74166 75p LM555	90p TBA753Q 200p 180p TBA800 80p 25p TBA810 100p	AF239 45p BC AF279 50p BC AU110 180p BC	338 18p BF258 2 348 20p BF259 2 461 35p BF271 2	5p TIP41A 70p 5p TIP41B 75p 5p TIP41B 75p	Mullard £1 25	each.
7447 50p 74170 100p LM710C 7447 50p 74170 100p LM710TC 7448 50p 74173 80p LM710D	40p 18A320 100p 5 69p TBA920Q 280p L 65p TCA270Q 220p	BA114 12p BC BA121 12p BC BA154 12p BC	516 35p BF274 3 517 35p BF324 3 547 12p BF336 3	0p TIP42A 80p 0p TIP42A 80p 5p TIP42B 85p	DECODER B	OARD
7450 12p 74174 60p LM723TC 7451 12p 74175 60p LM723DI 7453 12p 74176 50p LM723DI	05 40p TCA270S 220p L 40p TCA760 300p 120p TCA4500A 450p	BA157 15p BC BA173 15p BC BAX13 5p BC	547B 1310 BF337 33 548 1210 BF451 2 548C 1410 BF367 2	5p TIP2955 70p 5p TIP3055 50p	CONTAINING 18 × 74156 2 - 7	4155 2 × 7409
7454 12p /41// 50p LM741 7460 12p 74178 75p LM748 7470 25p 74179 120p LM1303N	20p IDA1008 300p 40p TDA1034 450p 100p TDA2002 300p	BAX16 5p BC BAW21 20p BC BB105 35p BC	549B 13p BF394 3 549C 14p BF458 5 557 13p BF458 5	5p TIS90 25p 0p IN914 5p	1 · 74180 1 ∕ 74 2 ∕ 60 WAY EDGE CONNE	150 1 · TIP32
7472 20p 74180 90p LM1458 7473 25p 74181 130p LM3080 7474 25p 74182 50p LM3900	100p IDA2020 300p 75p TL084 120p 55p XR320 250p	BB110 35p BC BC107 10p BC BC108 10p BC	557B 15p BF596 5 Y34 80p BF597 5 Y38 100p BF597 5	0p IN3754 20p 0p IN4001 5p 0p IN4002 5p	Few only left c unrepeatable b	of this bargain £3 50 each
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7481 85p 74190 70p MC1314F 7482 75p 74191 70p MC1315F 7482 75p 74192 60p MC1315F	190p XR2216 650p 230p XR2567 250p XR4136 150p	BC113 12p BC BC114 15p BC BC115 15p BC	Y59 250 BFW88 3 Y70 150 BFW88 3 Y71 200 BFW89 3	0p IN4006 6p 0p IN4007 7p 0p IN4148 4p	callers only) £ Bias Rejector	colls 50-100KHZ
7484 70p 74193 60p MK50596 7484 70p 74194 55p MM5314 7485 60p 74195 50p MM5316	380p XR4202 150p 380p XR4212 150p 480p XR4739 150p	BC116 15p BC BC117 18p BD BC118 12p BD	Y72 15p BFX29 2 115 45p BFX29 2 121 85p BFX34 2	5p 2N456A 90p 5p 2N929 20p 5p 2N930 20p	osp each	
7486 25p 74196 50p NE529K 7489 130p 74197 50p NE555 7490 25p 74198 100p	150p ZN414 100p 25p 95H90 700p	BC119 25p BD BC125 16p BD BC125B 20p BD	131 35p BFX38 2 132 35p BFX44 3 133 45p BFX48 4	5p 2N1302 25p 5p 2N1303 30p 5p 2N1305 30p	1MHz Crystals Push to Make 20p each.	Switches
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7494 70p 74 \$112 80p 7495 45p 7496 45p 7496 45p 0.47/2	CAPACITORS 5 7p 47/10 8p	BC137 16p BD BC138 30p BD BC140 30p BD	139 40p BFX88 2 140 40p BFY10 3 144 160p BFY18 10	5p 2N2219A 25p 0p 2N2483 30p 0p 2N2483 16p	4 Digit Phosp With A.M./P.I	hor Diode Dísplay M./Colon £5 00
7497 120p 7805 100p 1/16 74100 80p 7805 100p 1/25 74104 40p 7812 100p 1/25 74104 5 100p 1/25 1/50	7p 47/16 8p 7p 47/25 8p 7p 47/35 8p	BC141 30p BD BC142 30p BD BC143 30p BD	181 100p BFY19 10 182 100p BFY50 2 188 130p BFY51 20	0p 2N2907 20p 0p 2N3053 20p 0P 2N3054 50p	PRE SET POTS	MICRO BLOCK
74105 400 7815 100p 2.2/25 74107 25p 7818 100p 2.2/35 74108 100p 7818 100p 3.3/25	7p 47/50 9p 7p 100/10 8p 7p 100/16 8p	BC147 10p BD BC148 10p BD BC148C 14p BD	207 70p BFY52 20 220 65p BFY53 21 222 65p BFY90 12	0p 2N3055 60p 5p 2N3702 11p 5p 2N3703 12p	Vertical 50R—IM Ohm 81p Each	2102 250 Nano-Sec Static RAM (1024 x 1 BIT) £2 20 each
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2200/16 35p 4700/63 129p 10/16 2200/63 80p 4700/70 135p 10/25 2200/100 150p 10000/10 100p	7p 330/25 17p 7p 330/35 18p 7p 330/35 18p	BC167A 12p BD BC168 14p BD BC169 15p BD	609 80p BT100A 80 610 80p BU105 200 X32 200p BU133 180	0p 2N3866 95p 0p 2N3904 15p 0p 2N6027 50p	LÁ4 10-30KHZ 100p LA5 30-100KHZ	2112 450 Nano-Sec Static RAM (256 x 4
3300/30 50p 10000/25 150p 22/6V3 3300/63 90p 15000/15 150p 22/10 4700/25 50p 22000/25 200p	7p 470/10 14p 7p 470/25 19p	SPECIAL QU	ANTITY PRICES	2SD234 50p 2N5777 60p	100p LA7<10KHZ 100p	BIT) £2.50 each ★
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4007 14p 4054 100p POL 1 4009 30p 4055 130p 4011 12p 4056 120p 1000 P	F 5p 0-1 UF 6p	F.E.T. 10 for £2-30 100 for £20-00	GANGED POTS	BF195 BF196 BF197	1 AMP 50V 30p 1 AMP 100V 30p	8080 AN MPU (CPU)
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4019 40p 4081 12p 0.033 (4020 50p 4082 12p 0.047 (4022 50p 4093 70p	JF Sp 5°8 UF 40p JF Sp 10 uF 60p	MULLARD MODULES LP1152 100p	1M+1M LIN or LOG 2M+2M LIN or LOG	GN4A 125p GN9A 125p VX9181 70p	2 AMP 200V 55p	£2:00 each
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4029 30p 4383 70p 0·22/3 4030 30p 0·33/3 4032 80p 1 0·47/10	V 14p 4·7/35V 14p V 14p 6·8/6V3 14p	LP1181 400 p LP1400 400 p EP9000 280 p				
4043 60p + 0.68/3 4043 60p + 1.00/14	5V 14p 1.0/35V 14p 5V 14p 22/15V 21p 5V 14p 22/15V 21p	EP9001 280 p EP9002 280 p		PU	WE	
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MARCH 1979 VOLUME 55 NUMBER 3 ISSUE 865

Published by IPC Magazines Ltd., Westover House, West Quay Rd., POOLE, Dorset BH151JG

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Components are usually available from advertisers. A source will be suggested for difficult items.

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	"HINTS & TIPS FOR CONSTRUCTORS" — A special 8-page supplement on all aspects of construction
We re	gret that, owing to circumstances beyond our control, the size of

this issue of *Practical Wireless* has had to be reduced at short notice by eight pages. We apologise to readers for any disappointment caused.

Practical Wireless, March 1979

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NEW. CL22 aerial tuner which will match almost any receiver to almost any aerial at any frequency between 1.5 and 30 MHz. Six switch-ed ranges with fully tunable receiver and aerial matching capacitors. A worthwhile addition to any SWL station and an instant improvement in aerial matching problems. Price £15.75 inc. vat. P & P 66p. NEW. Bellsonic power supply giving fully regulated 12 V dc output at 3 amps continuous rating and 5 amps peak rating from 220-240 V ac mains incur

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Practical Wireless, March 1979

3

TRANSMITTER RECEIVER. MK.123. Very compact Army unit for use in range 2.5 to 20 Mc/s, receiver section 7 valves inc RF stage & BFO provides O/P for 4 K ohm phones, 3 bands with direct cal 2.5 to 5,5 to 10 & 10 to 20 Mc/s. Tx section 2.5 to 20 Mc/s in 3 ranges O/P 15/25 watts over range C.W. only, 2 valves Crystal Osc & 5B251M P.A. stage, will match into the following Aerial loads 25, 100, 500 & 1500 ohms as int tune up meter, reqs crystals type FT243 in range 2.5 to 10 Mc/s int morse key fitted with plug for ext key. Mains P.U. self contained unit for 115 or 200/250v AC overhaul size inc Rx, Tx & P.U. $30\times9\times14$ Cm weight 4 Kg. also supplied with ext invertor unit for 12v DC. Supplied in clean cond with 80 page handbook giving circ details etc (no details on 12v P.U.) £54.

TAPE RECORDERS. Made for use in language lab equipment 240v I/P uses BSR type TD.10 3 speed deck 5" spools, two chan transis amps with provision for playback & record can be used for stereo. P.U. and circ boards are mounted below tape deck approx size $12 \times 11 \times 7$ " intended to work phones. Supplied in clean condition may be less knobs & ind lamps some circ details supplied. No ext case. £13 also sim valve unit TD.2 deck £8.50.

U.H.F. CAVITIES. New spares for 2C39/7289 valves will tune over range 990/1040 Mc/s with int fittings circ supplied £6.50, also Rx section tunable preselector 1080/1130 Mc/s 4 section with 1N21 mixer diode for 60 Mc/s IF with circ £4.50.

ELECTROSTATIC VOLTMETERS range 0 to 15Kv AC or DC usable scale 3 to 15Kv complete in wood carrying case $8 \times 9 \times 6''$ f10.80.

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RECTIFIER UNITS ex-Army unit 200/250v I/P provides two DC O/Ps of 12v DC at 3 amps ea can be connected for 12v 6 amp or 24v 3 amp will do 4 amp okay for battery chargers good cond ± 10.80 .

INFRA RED LAMPS light & heat units sealed beam 115v 500 watts size 7" dia $4\frac{1}{2}$ " deep new American G.E. okay for paint drying etc two for £5·40.

V.H.F. TEST SET provides RF O/P over range 20 to 88 Mc/s in 4 bands, as int 2 Mc/s check, noise generator with 50 Ma meter, int pulse or C.W. O/P complete in case with cal charts & Ae rods, circ & notes. These req ext supplies of $250v \& 6.3 \pm 13$.

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PANEL METERS. 100 Ua scale 0 to 100 linear 2'' dia £3 also 1 Ma FSD special scale 2'' dia £1.30 both new.

FREQ METERS type BC221 125Kc to 20 Mc/s with int crystal check, with charts & book require 135v HT & $6.3v \pm 27$.

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Colour Brochure

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catalogue has been held up for a few weeks - since we have just been appointed as distributors for two of the most exciting ranges of radio components products yet : The Micrometals range of iron dust torroids cores and formers, and the OKI range of VLSI for digital frequency displays for receivers. We apologize for any inconvenience, but these two ranges are really worth the wait, and include some products you will find hard to believe, like the MSM5523 IC, an IC with less than ten external components that gives AM frequency readout to 1kHz from LW to 39.999MHz, FM frequency readout in 100kHz steps - (all usual IF offsets programmable by diodes), a 24 hour format clock with 12 hour display, independent on and off timers, time signals on the hours, stopwatch facility and a sleep timer. This costs £14 with its timebase crystal, and makes all that has gone before an expensive and time wasting excercise. Rather like the way the Intersi ICM7216 has revolutionized the instrument counter market. (See the OSTS ad.) And those of you familiar with Amidon and IG dust torroids, favoured in many new RF designs, will be pleased to know Ambit will be stocking a broad range of the Micrometals types for applications from EMI filters to RF PA stages. OK1 frequency counter ICs: details in cat2

for CA LEDs with RHDP such as FND507 £14 inc xtal for 3% digit LCD AM/FM with direct segment drive, no clock or timers £11 inc xtal MSM5523 MSM5525 Other types for fluorescent displays etc OA

Other new Semiconductor additions: KB4437 pilot cancel mox decod pilot cancel mpx decoder 4.35 muting stereo preamp 2.22 supercedes TDA2020 2.99 HiFi AM/FM 3.35 **KB4438** HA1370 TDA1090 TDA1090 HIFI AM/FM 3.35 TDA1220 low cost AM/FM 1.45 PRICES DOWN ON VMOS: as expected, this new technology in power transistors is getting cheaper. 120v comp pairs /100W for £10.00 Price reduction on CA3189Enow £2.20 New varicaps: to add to the biggest range.....

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A brief summary of some of our range of 1Cs: TDA1062/1.95; TDA1083/1.95; HA1197/E1.40 CA3123E/1.40; TBA651/E1.81; CA3089/1.94 HA1137/E2.20; MC1310/E2.20; HA1196/E3.95 KB4424/E2.75; KB4432/E2.53; SD600/E3.75 KB4412/E2.55; KB4413/E2.75; KB4417/E2.55 MC1495L/E5.86*; MC1406P/E1.25 LM381N/E1.81; LM1303/E0.99; ULN22838/ E1.00; LM380N/E1; TBA510AS/E1.09 TCA940E/E1.80; TDA2002/E1.95; ICL8038CC/E4.50*; NE5661/E3.50; NE567 E2.50*; NE5608/E3.50; NE56618/E3.50; NE5628/E3.50*; NE566A/E2.50*; SEE THE OSTS ADVERT FOR CMOS/TTL HEGULATORS, OPTO DISPLAYS, and other types of linear devices. A brief summary of some of our range

Some transistors for RF specifically: BF256LB/0.34; 40822/0.43*; 40823/0.51 * 40673/0.55* BF900/60* BF224/0.22; BF241/0.18; BF195/0.18; BF240/0.22; BF241/0.22; BF362/0.70; BF479/0.86; BF679S/0.70; BFY90/0.90*

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TERMS etc: CWO please, VAT on Ambit Items is generally 12%%, except where marked (*). Catalogue part 1:45p, part 2 50p all inclusive. Postage 25p per order, carriage on tuner kits £3. Phone Brentwood (0277) 216029/227050 9am-7pm. Callers welcome inc. Saturdays .

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That's not to say it doesn't look like HiFi - just that it doesn't look like the usual sort of bind to deal it does it took interint plat that it does to took interior shares and outperforms all British made HiFi tuners, and most imported ones too. Certainly at the price, there isn't one near it. But more than that, it looks superb . A small pic here would be an insult, so send an SAE for details on the kit that looks as if isn't. It's something else.

- Exceptionally high performance exceptionally straightforward asses Baseboard and plug-in construction. Future circuit developments will readily plug in, to keep the MkIII at the forefront of technical achievement Various options and module line-ups possible to enable an installment approx
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and now previewing the matching 60W/channel VMOS amplifier:

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In much the same as we have swept away the 'old technology' in frequency/timer In much the same way as we have swept away the 'old technology' in frequency/timer counters - with the OKI and Intersil single IC counters, we now offer a single IC "All Band" radio tuner. Don't confuse this one chip radio with things like the ZN414 - for this is a genuine superhet receiver with a mechanical AM IF filter, and ceramic IF filters for FM. The AM section employs a balanced input mixer section, covering all broadcast bands - plus a BFO and MOSFET product decetor for SSB/CW - though at this price, the tuner is not intended as a "communications receiver" - although we know of many lesser designs that make that claim. The AM sensitivity is nevertheless better than 5uV, and FM sensitivity is 1.2uV for 30dB S/N. As a multiband broadcast superhet receiver, it is a unique constructor project that fulfills the requests we very foruse the receiver, it is a unique constructor project that fulfills the requests we very foruse the receiver. project that fulfills the requests we very frequently get for a general coverage circuit that isn't over complicated. The set has CA3089E FM performance, with mute etc., and a PLL

Isn't over compinated. The set has choose if in performance, which must set a set as set as the set of the set An SAE for full details please. See the feature article in Practical Wireless (Dec/Jan)

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Practical Wireless, March 1979

The Sinclair PDM35. A personal <u>digital</u> multimeter for only £29.95



Now everyone can afford to own a digital multimeter

A digital multimeter used to mean an expensive, bulky piece of equipment.

The Sinclair PDM35 changes that. It's got all the functions and features you want in a digital multimeter, yet they're neatly packaged in a rugged but light pocket-size case, ready to go anywhere.

The Sinclair PDM35 gives you all the benefits of an ordinary digital multimeter – quick clear readings, high accuracy and resolution, high input impedence. Yet at £29.95 (+8% VAT), it costs less than you'd expect to pay for an analogue meter!

The Sinclair PDM35 is tailormade for anyone who needs to make rapid measurements. Development engineers, field service engineers, lab technicians, computer specialists, radio and electronic hobbyists will find it ideal.

With its rugged construction and battery operation, the PDM35 is perfectly suited for hand work in the field, while its angled display and optional AC power facility make it just as useful on the bench.

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Sharp, bright, easily read LED display, reading to ± 1.999 . Automatic polarity selection. Resolution of 1 mV and 0.1 nA (0.0001 μ A).

Direct reading of semiconductor forward voltages at 5 different currents. Resistance measured up to $20 M\Omega$. 1% of reading accuracy.

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The PDM35 will resolve 1 mV against around 10 mV for a comparable analogue meter – and resolution on current is over 1000 times greater.

The PDM35's DC input impedance of 10 M Ω is 50 times higher than a 20 k Ω /volt analogue meter on the 10 V range.

The PDM35 gives precise digital readings. So there's no need to interpret ambiguous scales, no parallax errors. There's no need to reverse leads for negative readings. There's no delicate meter movement to damage. And you can resolve current as low as 0.1 nA and measure transistor and diode junctions over 5 decades of current.

Technical specification

DC Volts (4 ranges) Range: 1 mV to 1000 V. Accuracy of reading $1.0\% \pm 1$ count. Note: 10 M Ω input impedance. **AC Volts (40 Hz-5 kHz)** Range: 1 V to 500 V. Accuracy of reading: $1.0\% \pm 2$ counts. **DC Current (6 ranges)** Range: 1 P A to 200 mA

Range: 1 nA to 200 mA. Accuracy of reading: $1.0\% \pm 1$ count. Note: Max. resolution 0.1 nA.

Resistance (5 ranges)

Range: 1Ω to 20 $M\Omega$. Accuracy of reading: $1.5\% \pm 1$ count. Note: Also provides 5 junction-test ranges.

Dimensions: 6 in x 3 in x 1½ in. **Weight:** 6½ oz. **Power supply:** 9 V battery or

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We'll mail your PDM35 by return! Sinclair Radionics Ltd, London Road, St Ives, Huntingdon, Cambs., PE17 4HJ, England. Regd No: 699483.

Send to: Sinclair Radionics Ltd, London Road, St Ives, Huntingdon, Cambs., PE17 4HJ. Please send me a Sinclair PDM35 @ £29.95 plus £2.40 (8% VAT) and 65p P&P, total £33.00.

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Practical Wireless, March 1979

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O those readers who have suffered delay and disappointment in obtaining their copies of *Practical Wireless* of recent months, we offer our sincere apologies. In some of the more remote parts of the UK, late publication and some hiccups in

distribution have combined to produce a particularly trying situation. We hope that these difficulties are now being overcome and that the magazine will be on sale by the specified date each month.

Another disappointment for some of our readers, is being unable to obtain a copy of an issue because every one has been snapped up, and their newsagent just cannot get any more. Worse still is picking up a casual copy of *PW* only to find that an interesting series has been running for some months and they can't get hold of the previous issues. We always do what we can to help someone in this predicament, and have recently increased the number of magazines put aside each month for our back numbers service (see page 1). Unfortunately though, many issues are now completely out of print, and only very limited stocks exist of some others. The table below summarises the situation as we went to press.

	1979	1978	1977	1976	1975	1974
JANUARY	\checkmark		\checkmark			
FEBRUARY	~	~			\checkmark	
MARCH		~		~		
APRIL			~		\checkmark	
MAY						
JUNE		~				
JULY						
AUGUST		~				\checkmark
SEPTEMBER		~				
OCTOBER		~			~	
NOVEMBER		~				
DECEMBER		~				

If you want to make sure of getting your copy of *PW* each month, either place a regular order with your newsagent, or take out a subscription—again, see page 1 for details.

Debbie Chapman, Copy Typist

A secretarial course which she completed immediately after her normal schooling at the Purbeck School, Wareham, led Debbie into her first job which, luckily for us, was with *PW*.

She was born in Kenya, where her father was serving in the Army, and most of her life so far has been spent on a "world trip" with her father and the rest of the family. Dad is an instructor with the 3rd Royal Horse Artillery at Bovington Camp, specialising in tanks. Mum also works for the Army, and Debbie is engaged to a Police Constable—she certainly maintains an interest in uniforms!

Among her other interests, which include swimming and looking after children, she quotes cooking as the most important. Future husband Martin seems to have it made in no uncertain terms—good food and a willing baby sitter while he's out working can't be bad.



Well done Erzsebet

The 1978 Girl Technician Engineer of the year is Mrs Erzsebet Kibble (see News, October 1978): the award was announced by The Secretary of State for Education and Science, The Rt Hon Mrs Shirley Williams, MP, at a meeting of The Institution of Electrical and Electronics Technician Engineers (IEETE) in London on Monday, 4 December.



Erzsebet, aged 25, is an Assistant Test Manager with Thorn-Ericsson Telecommunications (Sales) Ltd. Born in Hungary, Erzsebet is married to an Englishman and lives in Woking, Surrey. She joined her present employers in 1974 and gained the Higher National Certificate in Electrical and Electronic Engineering in 1976.

The Award, accompanied by a cheque for £250 and an inscribed rose bowl, is sponsored by The Caroline Haslett Memorial Trust and the IEETE. Its aim is to focus attention on electrical and electronic engineering as a worthwhile career for women. *IEETE, 2 Savoy Hill, London WC2R OBS. Tel:* 01-836 3357

New catalogues

Barrie Electronics Ltd. inform me that their latest catalogue is available. Although the 32 page catalogue lists a good deal of Barrie's stock, they have many other components, too numerous to list. They also invite readers to contact them if they need assistance with purchasing any difficult items.

The catalogue is obtainable for 15p, from: *Barrie Electronics Ltd., 3 The Minories, London EC3N 1BJ. Tel: 01-488 3316/7/8.*

Suhner Electronics, the UK subsidiary of the Swiss company Huber & Suhner, have recently issued 2 new catalogues covering just part of the range of connectors they manufacture.

1. High Voltage Connectors; this 16 page catalogue covers 3 different series of coaxial high voltage connectors with 50 ohm nominal impedance and bayonet coupling mechanisms. In all a total of 80 different connectors.

2. RF Connectors Series TNC; this comprehensive 28 page catalogue details Suhner's range of TNC medium-sized coaxial cable connectors.

Copies of these catalogues are available, free of charge, from: *Suhner Electronics Ltd., Telford Road, Bicester, Oxfordshire.*

Three new parts of the Mullard Technical Handbook are now available. They are: RF power devices—Book 1, Part 2— \pounds 2.00; Thyristors and triacs— Book 1, Part 5— \pounds 1.50; Loudspeakers, television assemblies and modules— Book 3, Part 5— \pounds 2.00.

The Mullard Technical Handbook is broken down into twenty-one different parts which can be purchased separately—the prices depend on the contents and include P&P.

The handbooks and a leaflet giving further details is available from: *Central Enquiry Handling Unit, Mullard Ltd., New Road, Mitcham, Surrey CR4 4XY.*

Operation Drake

Radio communications equipment, supplied by companies in the GEC-Marconi Electronics group, will play a big part in Operation Drake—the two year round-the-world voyage by parties of experienced explorers, scientists and 24 young explorers in the 150 ton brigantine *Eye of the Wind*, which left Plymouth on Wednesday, 8 November, to celebrate the 400th anniversary of Sir Francis Drake's circumnavigation of the globe.

Marconi Marine is providing one of its new 400 watt Transocean/Pacific s.s.b. radiotelephones to satisfy the vessel's requirements for m.f. and h.f. communications with shore-stations round the world. This equipment is powered by the vessel's a.c. mains supply.

In case of generator failure on board, Marconi Marine is also supplying a 24V d.c. operated Falcon II, a m.f./h.f. s.s.b. radiotelephone as a back-up set for the main equipment.

When working within 40 miles of the coast, the primary ship-to-shore

communications will be supplied by an Argonaut S v.h.f. radiotelephone.

For emergency purposes Marconi Marine is also to provide a Survivor II survival craft radio equipment.

Communications in the v.h.f./f.m. bands will be provided by three UK/VRC353 transceivers for the overland expeditions.

PW Gillingham

It has come to our notice that readers are having difficulty obtaining the 1 280MHz crystal for the PW Gillingham Short-wave Receiver Frequency Readout, published in the October 1978 issue.

A special arrangement has been made to supply these devices, at an inclusive price of £4.15, from *P. R. Gollege Electronics, Merriott, Somerset TA16 5NS. Tel: (0460) 73718.*

Project Index

A new index of electronic projects has been compiled by M. L. Scaife, Principal Librarian (Technical) of North Tyneside Libraries and Art Department.

The index provides a descriptive guide to over 2500 projects published between 1972 and 1977, in journals such as *Practical Wireless, Practical Electronics, Wireless World* and *Television Magazine.*

The project is not intended as a profit-making venture, as the following inclusive prices indicate: 1-2 copies— ± 1.50 , 3-6 copies— ± 1.40 , 7-10 copies— ± 1.35 , over 10 copies—special rates.

Copies of the index are obtainable from: *M. L. Scaife, Central Library, Northumberland Square, North Shields, Tyne and Wear NE30 1QU. Tel: (08945) 82811.*





OSO? GB2RN

The amateur radio station aboard HMS Belfast moored in the pool of London, between Tower Bridge and London Bridge, has been granted the use of the special callsign GB2RN when the ship is open to the public. Summer hours 1100 to 1800, winter hours 1100 to 1630, all British local time.

The station is interested in establishing schedules with other museum and special interest stations worldwide, these and other stations who would like to arrange skeds, please contact Don Walmsley G3HZL, 153 Worple Road, Isleworth, Middlesex TW7 7HT, England.

All h.f. bands from 1.8 to 28MHz are covered on c.w. or s.s.b.; it is hoped to have RTTY in the near future.

G4HMS will be operational outside of the stated hours.

LCDs

Two liquid crystal displays (l.c.d.s) will be available from Mullard in quantity early in the year. These devices join the Company's already established ranges of solid-state and plasma display devices. This announcement is the result of a long-term development programme which not only involved investigation into different techniques but also into basic materials in order to ensure that the displays would give continuing high-standard performance coupled with assured long-term reliability.

The technology selected for the Company's l.c.d.s was field-effect twisted-nematic. This technology has been in the market place for some time, but the Mullard l.c.d.s will incorporate the latest developments in this field. In particular, nematic liquid crystal materials have been developed to provide high standards of chemical stability and special precautions have been taken to eliminate chemical reaction between the different materials used. Also, to ensure satisfactory viewing, the l.c.d.s incorporate the results of a great deal of research into viewing angles, colour balance and the brightness ratio of the character to the background areas.

Many different types of l.c.d.s are in use today in watches, clocks, calculators and a great deal of instrumentation where their low power consumption, compared to that of l.e.d.s, enable them to be driven directly by m.o.s. devices.

The Mullard I.c.d.s were designed specifically for time displays and provide a 4-digit read-out with a colon between the second and third digits. Each digit is 12mm high and is formed from 7 segments. The first of these I.c.d.s, type LTC001R, is designed for use in the reflective mode; that is, it collects ambient light and reflects it back to the viewer. The second l.c.d., type LTC001T, is used in the transmission mode with a light source behind the display.

Mullard Limited, Mullard House, Torrington Place, London WC1E 7HD.

GB3WM

The Home Office have granted the special callsign GB3WM to the Wireless Museum at Arreton Manor, near Newport, Isle of Wight. Arreton Manor is the home of The Count and Countess Slade de Pomeroy, the Manor is open daily from Easter until November, and also by appointment during the winter months.

The Wireless Museum contains a unique collection of very early radio and television receivers, including a 30-line Baird Televisor with spinningdisc mechanical scanning.

It is now planned to build an exhibition short-wave experimental transmitting station, as used by radio amateurs before the last war, and display it side-by-side with a modern piece of equipment, showing a direct comparison between the old type of rack-mounted equipment and today's "state of the art".

All amateur bands will be used, with both c.w. and s.s.b. on the high frequencies and f.m. on two-metres (via GB3SN, the Alton, Hampshire repeater).

Official station operators will be Messrs. D. Byrne G3KPO (Museum Curator), A. R. Williams G3KSU, D. E. Denny G3ZQE, A. Wakeley G3MAD, R. W. Fisher G2DZN, K. B. Pearse G3MLC, F. D. Cawley G2GM, H. Childs G3IOW, L. Critchley G3EEL, D. Hoult G400 and W. Carter G2NJ.

A special QSL card will be printed and issued to all initial QSOs. Shortwave listeners are also invited to send in report cards, for which QSLs will be exchanged-all via the RSGB Bureau, or on receipt of an s.a.e.

D. Byrne G3KPO, The Wireless Museum, Arreton Manor, near Newport, Isle of Wight.



Practical Wireless, March 1979

Caller's opposite by appointment

Combridge 2'BT HDB







Practical Wireless, March 1979

So You Want to Pass the RAE?

A reprint of the complete series, including details of the new examination format being introduced in 1979, is now available. The reprint will cost 85p, including postage and packing to addresses within the United Kingdom.

Order your copy by completing and returning the coupon, together with your remittance, to IPC Magazines Ltd., Post Sales Department, Lavington House, 25 Lavington Street, London SE1 0PF. Please ensure that your name and address are clearly legible.

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Radio 3

Sir: While admiring the ingenuity of the BBC Engineering Department in replacing Radios 1, 2, and 4 transmitters, I must complain about the allocation of 1215kHz for Radio 3. This frequency is only fit for its original purpose—a back-up for l.w. transmissions in big cities. The map on page 41 of 18-24 Nov 1978 issue of *Radio Times* confirms this. The 1215kHz frequency is subject to hideous phase distortion here, and swamped by Radio Albania after 3 pm.

I am primarily concerned with conditions during daylight hours. Most people have v.h.f. at home, although that is difficult in some parts of this town, and v.h.f. car radio is ineffective here.

I am waiting for the howls of protest, when we have the first "Test Match Special" on m.w. only. I have had a long correspondence with the BBC and I am now hopeful that they will, after 23 November, try to improve Radio 3 outlets. Small relays could go on an international common frequency—or even a borrowed one, as at Cambridge (that is not on the Geneva list). What we urgently need is one more frequency to be used by alternative main stations as for Radios 1 and 2. We still have, as yet unused, an allocated frequency of 227kHz, although it may be liable to interference.

It is a scandal that the Home Office did not even ask for extra channels at Geneva (see Richard Last, page 12, *Daily Telegraph*, 13 November 1978). How did small countries like Belgium, Holland and Eire, manage to get new valuable long and medium wave allocations, all more efficient than the 648kHz, which is now lost to us, for the BBC European Service.

I hope that all your readers who enjoy Radio 3, especially in cars, as I have done for years, will make their views known to the BBC as soon as possible. Then, perhaps, some improvements will occur.

> Dr H. S. Brodribb St Leonards-on-Sea

Any offers

Sir: I have for sale approximately 150 copies of various electronic magazines, such as PW, PE, EE and WW. Some of the issues date back to 1946/7, if anyone is interested, I would accept any reasonable offer plus postage.

I also have a quantity of electronic bits and pieces (valves, transistors, transformers, motors etc.), for which all I require is the postage.

Mr. H. C. Leivers 4 Cavendish Place Gwaun Miskin Beddau CF38 2RP



In recent vears, Sound-to-Light (STL) units, in various forms, have become extremely popular for discos, pop concerts, parties, shop displays and so on. In its simplest form, a STL unit consists of three of more CR filters fed from the loudspeaker terminals of an audio amplifier, feeding transistors which trigger triacs (or thyristors) via pulse transformers. Each channel has its own level control, and there may be a "master" control as well. This form of STL has several drawbacks; the sensitivity is low (an audio signal of 2-3V may be required) and continual adjustment is required to compensate for changes in volume and tone. The apparent separation between channels is poor, due to the simple filters used: the unit generates interference spikes on the mains, since the triacs may be triggered at any point on the mains cycle, and it uses several heavy and/or expensive transformers.

The STL unit to be described overcomes all of these disadvantages; it uses interference-free "zero-voltage switching", so that the triacs are triggered only when the

voltage across them is very low; it has active filters for excellent separation between channels and automatic gain control (a.g.c.) amplifiers to eliminate manual adjustments; the sensitivity is approximately 700mV to control 1kW of lights per channel, and it uses only one inexpensive transformer. Due to the virtual elimination of transformers, this design is little more expensive to construct than the conventional version described above.

The Circuit

As can be seen from the simplified block diagram (Fig. 1), the circuit comprises several blocks each with a welldefined action. Each block is identified in the circuit diagram (Fig. 2). The signal from the loudspeaker circuit of an audio amplifier is fed via transformer T1. This provides isolation between the amplifier and the remainder of the circuit, which is at **mains line voltage.** Consequently, **extreme caution** is required when constructing the circuit.





Fig. 2: Complete circuit diagram (excluding options)

Note that T1 is actually a miniature mains transformer, which has good isolation and adequate frequency response for this application. The audio signal then enters an a.g.c. amplifier which allows the STL to operate with inputs from 700mV r.m.s. $(0.12W/4\Omega \text{ to } 25V (150W/4\Omega)$. This consists of a voltage controlled attenuator R16/Tr3, followed by a X30 amplifier IC2a, with a control voltage derived from the rectifier circuit D6/D7.

If more sensitivity is required, R13/R14 may be reduced to, say, $1k\Omega$ 1W, but this loads the audio amplifier somewhat more, and increases the possibility of noise affecting the lights. If a high voltage source is used (e.g., 100V lines) R13/14 should be increased to $15k\Omega$ 1W. Note that a stereo input is provided; if mono only is required, both inputs should be tied together, alternatively a mono jack plug in the (stereo) jack will work with slightly reduced sensitivity.

After processing in the a.g.c. amplifier, the audio signal is fed to three active filters: low-pass (bass, IC2b), bandpass (midrange, IC2c) and high-pass (treble, IC2d). These have their cut-off frequencies chosen to give subjectively good separation between channels with most types of music from a variety of sources from a portable transistor radio to powerful discotheque systems. The output of each filter is connected to one of three identical detector/a.g.c. circuits constructed around a quad comparator integrated circuit.

For simplicity, consider the circuit connected to the low-pass filter (IC1b). If there is no audio signal, the comparator output will be low due to the bias voltage from D10 and R32, and no trigger pulses will be applied to the triac. Now, if a steady audio tone is applied, capacitor C14 will rapidly charge via D9 and the comparator output will go high. This allows trigger pulses, timed to coincide with the points in the mains cycle where the voltage is zero, to be applied to the triac. Thus, the triac will turn on when the voltage across it is small and the appropriate lights will come on without any interference being generated. However, C15 slowly charges via D8 and eventually the voltage across it exceeds that on C14. The comparator switches again, turning the lights off. With more complicated dynamics in the audio signal the situation is more complex, but it can be seen that this circuit provides







Fig. 3: (a) (top) Addition of pushbutton and switched override facilities. (b) Half/full power switching

A rear view of the prototype, showing input jack, fuses and eight-way output socket

voltage" pulses from the pulse generator circuit Tr1/IC1a. The mains input is clipped by D3, and an inverted squarewave is produced at Tr1 collector. The positivegoing edges of both are differentiated by CR networks, and trigger a comparator IC1a providing short positivegoing pulses at Tr2 collector. The positive supply for the whole unit is provided by D2 and C1.

Circuit Modifications

The circuit described so far is capable of forming a simple but effective "no frills" STL unit, with no external controls whatsoever! If, like the writer, you prefer more facilities, several additions can be made, some of which are provided for on the printed circuit board described later.

First, a mains on-off switch can be provided—useful for quick changes between lighting effects. Switches can be provided (Fig. 3(a)) to override the triacs on each channel, switching the lights permanently on. If the switches have built-in neon indicators, these can be connected to provide monitoring of the light display. These switches must be capable of carrying the full load current. However, low power switches or push-buttons may be used with a simple diode gate (also shown in Fig. 3(a)), providing fast, noisefree switching. This method is particularly suitable for push-buttons, allowing the lights to be rapidly flashed under manual control. These additions are included in the version described in the constructional details and positions for the extra components required are included on the p.c.b. Other improvements include the addition of a half/full power switch (Fig. 3(b)). This short-circuits Tr1, so that only half of the differentiated pulses reach the comparator. Thus, the triacs are only triggered on positive half-cycles, so the lights will be less bright. In a similar manner, an audio on/off switch can be added, allowing the override push-buttons to be used alone. This switch short-circuits R15, preventing audio signals from reaching the a.g.c. amplifier.

More ambitious modifications include making presets VR1-VR3 front panel controls (for the "knobs with everything" man!) or even providing controls with a switched "automatic" position. If this is tried, the controls **must** be well insulated. Also, a sequencer facility can be added, using transistors in place of the push-buttons in Fig. 3(a), or a dimmer control can be added using a 555 oscillator. These modifications are left to the more experienced constructor.

Construction

WARNING: Most of the circuit board and wiring is at MAINS LIVE VOLTAGE and EXTREME CAUTION is necessary when testing or adjusting the circuit. The use of an Epoxy Glass Fibre p.c.b. is strongly recommended; Veroboard, etc., is NOT suitable, as it has insufficient intertrack voltage rating.



Fig. 4 (above): Printed circuit board track pattern, shown full size.







Internal view of the prototype unit 🚄

The p.c.b. track pattern is shown in Fig. 4 and the component layout in Fig. 5. Constructors should have little trouble in assembling the board provided the specified components are used. The usual precautions against the effects of static electricity should be taken when handling the two c.m.o.s. integrated circuits. It is safest to use sockets as specified in the components list, and insert the i.c.s only when the remainder of the components have been fitted to the p.c.b. Note that the wirewound resistor R1 gets hot in operation and should be mounted away from the board. Also, the triacs may require larger heatsinks if the unit is constructed in a small, enclosed box. If a common heatsink is used, mica washers and insulating bushes will be required. If the push-button override facility is not required (this is shown on the layout diagram), D_A , D_{B} and R_{A} (three of each) may be omitted and D_{A} replaced by a wire link. Finally, the p.c.b. has extra holes to allow the presets to be replaced by fixed resistors when the optimum values have been found.

The wiring of the unit is not critical and any convenient layout may be used. For guidance, the wiring used in the writer's unit is shown in Fig. 6, drawn flat for clarity. All mains wiring should, of course, use thick wire, say 40/0.2mm, remembering that some leads can carry up to 12A! Note that diode D1 is mounted off the board and resistors R13–R15 are mounted on the input jack JK1. Any convenient case may be used, but the circuit must be securely protected from prying fingers. All metal parts except the heatsinks must be efficiently earthed. The writer's unit was constructed on a metal chassis in a wooden sleeve as shown in the photographs, but alternative construction methods could easily be used.

Fig. 6: Wiring diagram of a unit incorporating options of Fig. 3(a)



★ components

Resistors			Semiconductors
4W 5%			Integrated circuits
220Ω	3	R35,48,59	MUJJUFF 1 102 (OF LIVIJJUUN)
🐁 1kΩ	1	R3	MC3302P 1 ICT (07 LW3302W)
2.2kΩ	4	R11, 36, 49, 60	Transistors
3/3kΩ	1	R23	BC182L 4 Tr1, 4, 5, 6
4-7kΩ	4	R10, 34, 47, 58	BC212L 1 Tr2
: 10kΩ	10.	R2, 4, 7, 12, 19, 20, 28	2N5245 1 Tr3
	10.00	33, 42, 53	Diodes
22kQ	Contraction in the	R15	BZX61C15V 1 1 D3 (or similar 15V 1W Zener)
33kΩ	and the second	R17	BZY88C3V3 1 D10 (or similar 3.3V 400mW
47kQ	4	R21-31,45,56	Zener
100kΩ	5	R8, 29, 43, 50, 54	0A91 2 D6.7
- 1.50kΩ	7	R5, 6, 30, 40, 41, 44, 55	1N4004 2 D1.2
220k0	2	R16,25	IN4148 8 D4 5 8 9 11 12 13 14
330k0	1	R39	Transa
1M0	5	R22, 24, 26, 37, 51	T28000 2 CCD1 2 DU AUTON (0) C
2·2MΩ	8	R9, 18, 27, 32, 38, 46, 52, 57	triacs)
1W 5%	1978. 24		Miscellaneous
2.•2kΩ	2	R13, 14 (see text)	T1 Min, mains Transformer 12-0-12V secondary
1014/ 50			FS1-3 3A ceramic quick-blow fuses, with panel-
4.740	Star 1	81	mounting holders: JK1 6mm (4in) stereo lack (see
Potentiom	eters	e e e e e e e e e e e e e e e e e e e	text); SK1 8-pin socket, Bulgin P552 (plus mating plug P551); Mains on/off switch, d.p.s.t. (if required); Soldercon socket strips to mount IC1 and IC2; Heat-
Min. horizo	ntal pres	sets	sinks for CSR1-3. Printed circuit board. Box to suit
47kΩ	3	VR1,2,3	
		the standard	OPTIONAL EXTRA FACILITIES 1—Push-button control
Conselle			Resistors
50V plate	pramine		±₩5%
32 prace.t	1	C8'	4.7kΩ 3 R _A (x 3)
47005	2	C21 22 23	
10F	3	C17-18	Diodes
2.205	5	C11,12	1N4148 6 D _A , D _B (× 3)
A.7.5	2	C2 3	
+ (AUL)	52 L		Switches
1.6V tantal	um beac	C5	Normally open push-buttons s.p.s.t. 3
2.2HP	.0	C9 10-13 14 16 19-24-25	
10.5	0	C15-20.26	2-Switched override
10µF	3.	C6 7	Switches
Archt.	10 4		lituminated mains rocker s.p.s.t. 3
35V tantal	um beac		
1μF			3-mait/full power switch
25V electro	olytic		Switch Towards and the
1000µF	1	C1 C1	ruggie s.p.s.t.
M. M. There and the state	the statement of	The second s	

Testing and Setting Up

After construction is completed, all wiring should be double-checked and all presets set to their central position. Connect a set of lights and switch on. The lights may flicker briefly then they should remain off. The pushbutton switches (if fitted) should be tested at this point; this checks the operation of the triacs. If all seems well, switch off and connect an audio source. Setting up is best done with a moderately large input signal (2-3V r.m.s.). Switch on again and the lights should show some response. The presets should be adjusted, with an insulated tool, to give the best results with various types of music; this need only take 15 minutes or so. Finally, vary the volume over a wide range and check that the display is similar at different volume levels. The a.g.c. amplifier takes a second or so to "catch up" after rapid changes in volume, so allow a little time between adjustments, etc.

Any faults found are likely to be due to incorrect assembly (e.g., short-circuits on the board, reversed tantalum capacitors), faulty semiconductors or "dry" joints. Note that it is no use trying to test the unit using only the neon indicators in the override switches (where fitted), instead of external lamp loads. The neons do not draw sufficient current to hold the triacs on after they have been triggered.

In cases where there is an exceptional amount of electrical noise on the mains supply from other sources, the lights may flicker even when no audio signal is present. If this is a problem, a 0.1μ F 1000V capacitor across the mains supply should cure it (see Fig. 6); alternatively, or in extreme cases additionally, the sensitivity may be reduced slightly by increasing the value of R17.

The unit described has been very reliable in use, and several prototypes have been constructed. With care, constructors will have a Sound-to-Light unit of professional standard, capable of long service.

Some original circuit ideas provided by our readers. These designs have not been proved by us, and we cannot therefore guarantee their effectiveness. They should at least provide a basis for experimentation.

Why not send us your idea? If it is published, you will receive payment according to its merits. Articles submitted should follow the usual style of PW in circuit diagrams and the use of abbreviations. Diagrams should be clearly drawn on separate sheets, not included in the text.

Each idea should be accompanied by a declaration that it is the original work of the person submitting it, and that it has not been accepted for publication elsewhere.

STEP TONE GENERATOR



This circuit produces a sound which relates to defined increments stepping up and down a musical scale. No shift register or counter is required to produce each step, and with the exception of the peripheral components the circuit consists of only two 555 timers, an f.e.t., and a unijunction transistor.

One 555 is used as an astable driving a loudspeaker, with the other functioning as a triangular waveform generator providing a control ramp for feeding back into the astable.

Increments are provided by spikes from the unijunction at intervals of $100\mu s$, which are impressed onto the triangular waveform. The output of the astable is frequency modulated at the same rate, and as the triangular waveform changes, a series of loud beats are produced, running up and down the musical scale. The f.e.t. acts as a buffer to prevent loading of C1. This circuit could be modified to become a "blind man's d.v.m." or a relatively sophisticated voltage to frequency noise scrambler.

> D. Brown GM8FFH, South Hawthornhill, Dumbarton.

CAR CASSETTE POWER SUPPLY

This simple circuit uses a 7805, which is a compact device used mainly with TTL i.c.s and fed with an ample input of at least 8V will produce a stabilised output of 5V at up to 1 amp, provided it is mounted on a heat sink. The minimum of smoothing is required, and the overall circuit will provide 5.6V to power a cassette player from a car battery.

Only a handful of components are needed and R1 was selected such that no voltage drop occurred at maximum volume and of course, the voltage never rises beyond 5.6V without a load. There is, therefore, little chance of the device becoming faulty and damaging the cassette recorder. If a higher output voltage is needed, another diode





can be connected in series with D1 giving 6.2V but for safety, the existing output is sufficient for most applications.

The output can either be connected directly to the battery connectors in the recorder or alternatively a low voltage power plug can be used to connect externally to the recorder if a socket is fitted.

The unit will work with cars using either positive or negative earth but if a metal box is used as a case, care must be taken to prevent it coming into contact with any part of the car body, especially if a positive earth is used.

> M. Burrell, Halstead, Essex.

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<section-header>

VHF/UHF FOLDED COLINEAR AERIAL ARRAY

An omni-directional, vertically polarised design, giving 3dB gain over a dipole. Details are given for both 2 metre and 70cm versions



Practical Wireless, March 1979

THE CASSETTE TAPE MEDIUM

Catholifeon England

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FILEN

The development of the Compact Cassette system from its inception to the present day is reviewed by Gordon J. King

M. TOOLEY & D. WHITFIELD

After completing the assembly of the receiver as shown in Fig. 5, a careful visual check should be carried out on all wiring, the unit connected to a 12V d.c. supply and the supply current measured. This should be approximately 75mA in the absence of a signal. If a signal generator is available, this should be connected to the aerial socket via a suitable attenuator, and set to 455kHz (\pm 1kHz) with internal modulation. A tone should be heard when the receiver is switched to give a.m. reception, and it should be possible to peak IFT1 and the filter IFT2 for maximum output. If necessary, increase the attenuation between the signal generator and the receiver progressively as the i.f. circuits are peaked. This alignment is best done with a relatively weak signal since the a.g.c. action will, to some extent, mask changes in the output level for strong input signals.

An input to the receiver of about 1mV at 455kHz will be all that is required in order to produce a strong audio output signal if the i.f. stages are working correctly. If a signal generator is not available, the i.f. stages may be aligned by tuning the receiver for a weak signal and then adjusting IFT1 and the filter for maximum output. In either case, the alignment process is greatly simplified by reference to the signal strength meter. To adjust the meter, VR4 is first set to mid-position and, with the aerial disconnected, VR5 is then adjusted to zero the meter. VR4 may later be set to calibrate the meter. Constructors should note, however, that there may be a slight interaction between these two controls and the setting-up procedure may have to be repeated whenever the calibration is being altered.

After completing the i.f. alignment, the r.f. tuned circuits of L1 and L2 should be aligned. The signal generator should be set to 1.45MHz with internal modulation and VC1 set for maximum capacitance (i.e., vanes fully meshed). VC2 should be set to mid-position and the core of L2 adjusted until a strong signal is heard. The core of L1 is then peaked for maximum signal. The signal generator is then set to 3.5MHz and VC1 adjusted to near minimum capacitance (i.e., vanes almost fully open) until a strong signal is again heard. The core of L1 should again be peaked for maximum signal but the core of L2 should be left alone. If, however, the 3.5MHz signal is not located, the core of L2 may be adjusted until a signal is heard at minimum capacitance. This establishes the high frequency end of the receiver's tuning range. There will be a noticeable difference in the settings of the core of L1 at 1.45MHz and 3.5MHz, due to the "tracking" error. With VC2 still set to mid-position, the signal generator is tuned to 2.5MHz and a strong signal located at approximately mid-position of VC1 (i.e., middle of the tuning range). The core of L1 is peaked for maximum, when its position should be roughly mid-way between the 1.45MHz and 3.5MHz positions.

If a signal generator is not available, signals on known frequencies can be used as an aid to alignment. These could include broadcast signals at around 1.5MHz, coastal radio stations on frequencies around 2MHz, and the 2.5MHz standard frequency and time signal. The vernier dial can be calibrated by first removing the scale, painting it with a matt-finish white or silver paint, and then using dry transfers to provide markings every 100kHz from 1.5MHz to 3.5MHz. Alternatively, a somewhat simpler solution is to draw a calibration curve for the receiver and to refer to this whenever exact frequency readout is required. A typical calibration curve is shown in Fig. 6.

BFO

In order to align the b.f.o. transformer, IFT4, a strong carrier, preferably with no modulation, should be selected. With the receiver switched for a.m. reception, VC1 should be carefully tuned to provide maximum indication on the signal strength meter. The mode switch, S1, is then set to the s.s.b./c.w. position and, with VR3 set to mid-position, the core of IFT4 is adjusted for zero-beat.

A beat note may not, in some cases, immediately be heard since the heterodyne produced may lie outside the audible range. If this is the case, adjustment of the core of IFT4 should readily produce an audible note. The quadrature detector tuned circuit, IFT3, is best adjusted using a known f.m. signal of preferably 3kHz peak devia-


tion. The receiver is switched to f.m. and the core of IFT3 is then adjusted for maximum undistorted audio output.

Note that the recovered audio level may be significantly less on f.m. than for a.m. The volume control can be adjusted accordingly to compensate for this reduction in level. This completes the alignment and calibration of the receiver, which is now ready for "air testing".

Using the Receiver

The receiver is simple to use and automatic gain control is provided for a.m. reception, but this may be switched out if required when c.w. or s.s.b. signals are being received. This facility will be found to be most useful when strong signals are concerned; the r.f. gain is then simply

backed off for best results. The facility to receive f.m. (which is very rarely used at h.f.) was thought to be a useful addition to a receiver which can potentially be used as a tuneable i.f. in conjunction with a v.h.f. converter. The f.m. facility provides for the reception of narrow-band f.m. signals of typically not more than 3kHz peak deviation. Limiting action is automatically provided for strong signals. The f.m. detector will, however, not perform well with wide-band f.m. signals and, should this be an important consideration, constructors are advised to replace the i.f. filter IFT2 with a comparable type having a wider bandwidth (e.g., the CFT455B which has a nominal 8kHz bandwidth and is pin-compatible).

The receiver is eminently suitable for either fixed or portable use. In each case the receiver will benefit greatly from the use of a properly designed aerial system and tuning unit or preselector. This ensures optimum performance and also helps to reject signals on the image channel. These sometimes appear as breakthrough of strong h.f. signals spaced by twice the intermediate frequency (i.e., 910kHz) above the wanted signal. A particular problem associated with sensitive low- and medium-frequency receivers is that of line-timebase interference from nearby television receivers. There is often little that can be done to prevent this nuisance, but in exceptional circumstances it may be necessary to site the aerial some distance away from dwellings and to employ a buried coaxial feeder with its screen earthed at each end.

For mobile use, a simple base-loaded whip aerial should be adequate although, here again, some form of preselector or tuning unit will give vastly improved results.



Fig. 6: Tuning calibration curve, used as an aid to alignment



Aerial Matching

The additional tuned circuit shown in Fig. 7 will provide an extra 20dB of image channel rejection. The variable capacitor VC should have a maximum value of about 500pF. L1 consists of 100 turns of 36 s.w.g. enamelled copper wire wound in two layers on a 4.8mm diameter former fitted with a dust core, base and screening can. L2 consists of 10 turns of 30 s.w.g. over-wound on L1. C_x can be 47pF for short aerials or 15pF for a long-wire aerial.

If desired, this form of circuit can be modified to use a ferrite-rod aerial, as shown in Fig. 8, though it should be remembered that such an aerial is directional. Here, VC should have a maximum value of 208pF. L1 is 22 turns of 30 s.w.g. enamelled copper wire wound at one end of a 180mm ferrite rod (see Fig. 9), and L2 is 5 turns of 30 s.w.g. enamelled copper wire over-wound on L1. The ferrite rod should be mounted on pillars about 50mm long, either above or to the rear of the receiver.

An aerial matching unit provides an excellent method of improving the r.f. selectivity of the receiver, thus adding considerably to the image channel rejection, and also acts as an impedance match. In the circuit of Fig. 10, Tr1 is operated as an emitter follower which exhibits a voltage gain of slightly less than unity with a high input impedance and a low output impedance. The circuit can be constructed on a small piece of Veroboard or matrix board. Component details for the input tuned circuit are shown in the table.

Tuned Circuit Details for the Aerial Matching Unit

				A CONTRACT NO. NO. NO. MARK MARK
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<u> </u>	स्कृत्य क	and the second	v ∀ *',	e down wrober er fw
	∂5 turns:	22 turns	208pF	Not recommended
2.34	30 SWG	30 SWG	$\sim z_{\rm e}$	for use with external
	wound on	ferrite rod		aerials, directional
1	199 - 18 - 18 - 18 - 18 - 18 - 18 - 18 -	ant a start a		
,∦r.,	10 turne	100 turns	3650F	Recommended for
xé	30 SWG	36 SWG	or	use with external
્રહ્ય	wound on	4.8mm dia.	500pF	aerials
l ac	former wi	th dust core	t the last	化甘油 医病试验
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\star components

POWER SU Resistors	IPP	LYU	NIT
↓W 10% carbon			`
470Ω	2	R1,	2
1kΩ	:1 .	R5	
4·7kΩ	2	R3,	4
Variable Resist	ors	÷	·
Min. horizontal pi	rese 1	t 1	
10842	, , , ,	V111	
Capacitors			, ```
25V electrolytic		~~	 .
10µF	1	C2	,
470μF	. 1	C1	
Semiconductor	'S .		-
Diodes			
1N4001	4	D1,	2, 4, 5
BZY88C6V2	1	, D3	,
Transistors			
BC548	1	Tr2	-
BD131	1	Tr1	
Miscellaneous T1 240V primary FS1 1A 20mm fu Printed circuit bo	, 2 > use, ' ard.	< 15V with p Heat-	secondaries, 6VA anel holder sink for Tr1
WRM116		1	5
		and the second second	Fig. 9: Details of the ferrite rod aerial
All I			
			Fig. 10 (below) the circuit diagram of the aerial matching unit
WRM117			R4 270



+20.6\ , +13∙6V D1 A 240 1N40C Tr1 \ BD131 To S3 a 1N400 R4 0 D4 4.7k 1N4001 1N4001 B) VR1 С1 470,⊔F 10 12V iBattery ≤R3 ≤4·7k supply 470 D3 C2 BZY88 10_/uF C6V2 To S3b EC WAD285 +6.8/

Fig. 11: The circuit diagram of the mains power supply unit



Power Supply

The receiver will give many hours of operation from its internal battery supply which consists of eight U11 cells or equivalent. However, a mains power supply is a very useful addition to any receiver which is likely to be used for any length of time at a fixed location. The circuit diagram of a suitable supply is given in Fig. 11, with p.c.b.

Practical Wireless, March 1979

track pattern and component layout shown in Figs. 12 and 13 respectively.

A centre-tapped mains transformer T1, and a full-wave rectifier (diodes D1 and D2) are used to build up a d.c. voltage across the reservoir capacitor C1. The smoothed output is stabilised by the series regulator transistor Tr1, driven by a d.c. amplifier based on Tr2. Control of the output voltage from the unit is achieved by supplying the base of Tr2 from an adjustable tap on the potential divider



Copper track pattern and layout of the power supply unit shown full size

formed by R4, VR1 and R3. Zener diode D3 provides an emitter-reference potential for Tr2. Changeover from a.c. mains to battery power is automatic in the event of a supply failure. This is achieved by the steering diodes D4 and D5. When the mains supply is present, the potential at point "A" will exceed that at point "B" (see Fig. 11). This will reverse-bias D4 and forward-bias D5, effectively open-circuiting the battery supply. When the mains supply is removed, either by accident or design, D4 will be forward-biased and D5 reverse-biased, supplying battery power to the receiver. The power supply is straightforward to build and only one adjustment is necessary. Before connecting the output of the supply to the receiver board, the voltage at point "A" should be set to +13V by adjustment of VR1. In case of any problems, typical test voltages as measured using a $20k\Omega/V$ multimeter on the 25V range are indicated on the circuit diagram.

The a.c. mains plug should be removed from the supply when the receiver is not required for operation over an extended period, since the mains unit remains working independently of the front panel on/off switch.

THE **G**'VINTON' **Stereo Amplifier** Part 1

E.A.RULE

The Winton amplifier has been designed primarily for the home constructor who would like a high fidelity amplifier equal to the best commercial design available, but at a price that enables it to be built within a sensible domestic budget, the total outlay being in the region of $\pounds110$ for the complete project.

All the components are available and no special test equipment is needed for testing or setting up. The amplifier uses the very latest techniques and is capable of a standard of performance at least equal to the best commercial designs available with similar power ratings. In some aspects of its design it will outperform other amplifiers costing very much more.

Power f.e.t.s are used in the output stage and allow a wider power-bandwidth response with lower distortion than could be obtained if conventional bipolar output transistors were used. In the control unit section, bi-f.e.t. op.amps are used. These have an improved slew rate over the more common types, as well as much lower distortion. In the disc (magnetic) pre-amplifier, a three stage circuit using ultra low noise transistors enables noise figures of -68 dB unweighted relative to 50W output to be obtained, with distortion so low that the input (normally 3mV) has to be increased to 140mV (+33dB) before the distortion even reaches 0.1%. A full circuit description will be given later.

On the front panel the Winton amplifier has controls for Volume, Balance, Bass, Treble and push buttons for selecting, Disc, Tuner, AUX 1, AUX 2, Tape monitor, Mono, h.f. and l.f. filters. Plus switches for mains on/off and loudspeaker or phones.

The headphone socket is also mounted on the front panel. DIN sockets are fitted on the rear panel for all inputs, terminals for the loudspeaker connections, an earth terminal, and fuse for the mains and each loudspeaker output. The heat sinks for the power f.e.t.s are also mounted on the rear panel.

Power f.e.t.s

Until recently bipolar transistors have been used in almost every Hi-Fi available (the exceptions using valves). Bipolar transistors require a wide area of Safe Operation to achieve reliability and a large gain-bandwidth product so that large amounts of negative feedback can be used at the higher audio frequencies to reduce the distortion. They also have a positive temperature coefficient which means that any increase in transistor temperature causes an increase in transistor collector current which causes the temperature to increase further, and so on requiring careful design to avoid thermal runaway. Also, as bipolar transistors are minority carrier devices, they suffer from storage effects at high frequencies which can cause a most objectionable distortion, which may well account for the so called "transistor sound".

In the light of all these problems a considerable amount of money and research has gone into looking for a better device. One result of this has been the power f.e.t.s developed by Hitachi Ltd., of Tokyo, Japan. Some advantages over conventional transistors are:

- 1) Good frequency response because of fast carrier speed.
- 2) No storage effect.
- 3) Negative temperature coefficients, so no risk of thermal runaway.
- 4) No secondary breakdown.
- 5) High input impedance and high gain.

The Winton amplifier uses a complementary pair of Hitachi power f.e.t.s Type 2SJ48 (*p*-channel) and 2SK133 (*n*-channel). These devices have a maximum dissipation rating of 100W each, so when used in an amplifier of 50W output, each device is only dissipating about 25W, or just ticking over! In fact combined with a 120V drain to source breakdown voltage and a drain current of 7A they are almost indestructible when used in the Winton, which











Fig. 1: Circuit diagram of the PW "Winton" stereo amplifier, showing Channel "a" (left) only. The following components are shared between the two channels: R12–15, 37, 38, 62–64; C2, 3, 8–11, 24, 40–45; Tr13, 14; IC1–3; D1, 2, 8–12; VR3; T1; F2; S9. The remainder are duplicated for the other channel

★ specification

Power Output

Continuous sine wave power, both channels driven 50 + 50W into 8Ω

Power Bandwidth (power amp. only) -1dB 15Hz -100kHz

Frequency Response (power amp. only)-0.5dB10Hz - 40kHz-3dB5Hz - 150kHz

Harmonic Distortion (50 + 50W) 1kHz 0.013% 10kHz 0.015% 20kHz 0.018% 100Hz 0.011% 20Hz 0.019% No significant increase at lower powers

Intermodulation Distortion (28V pk into 8 Ω) $f_1 = 15$ kHz $2f_1 - f_2 0.005\%$ $f_2 = 16$ kHz $2f_2 - f_1 0.004\%$

 $\begin{array}{c} \text{i} \text{kHz} & 2f_2 - f_1 \ 0.004\% \\ f_2 - f_1 \ 0.003\% \end{array}$

Damping Factor

20Hz – 1kHz 80 20kHz 20

Rise Time (power amp. only) 2µs

Slew Rate (power amp. only) 26V/µs

Stability - Unconditional

Sensitivity for 50W

Disc 3mV Tuner 100mV AUX 100mV Tape 100mV

Input Impedance

Disc 47kΩ Tuner 100kΩ AUX 100kΩ Tape 100kΩ

Maximum Input for 0.1% t.h.d. Disc 140mV Tuner 4V AUX 4V

Frequency Response ± 0.5 dB Disc RIAA Tuner 20Hz–20kHz AUX 20Hz–20kHz Tape 10Hz–40kHz



Plot of Harmonic Distortion components generated when handling a 1kHz signal dissipating 50W in an 8Ω load. The 2nd harmonic is at -82dB and the 3rd harmonic -86dB relative to the fundamental



The upper trace shows the PW ''Winton'' output when handling a 1kHz signal dissipating 50W in an 8Ω load. Note the complete absence of crossover distortion

The lower trace shows the residual harmonic content of the above signal, after the fundamental had been removed by the distortion meter filter. The total harmonic distortion is about -78dB, but is masked by noise



Plot of Intermodulation Distortion components generated when driven to full power into an 8Ω load by two equal-amplitude signals, $f_1 = 15$ kHz, $f_2 =$ 16kHz. Component $2f_1 - f_2$ is at -86dB, $2f_2 - f_1$ at -88dB, and $f_2 - f_1$ estimated at about -90dB

This test, using these frequencies, is very critical of the power-bandwidth capability of an amplifier, and is seldom given in specifications due to the generally poor figures obtained

Hum & Noise

Unweighted* with reference to rated sensitivity Disc-70dB Tuner -75dB AUX-75dB Tape -75dB

Hum & Noise

Disc input with reference to 10mV input Unweighted 80.5dB

Cross Talk** -48dB

Tape Output 100mV AUX 1 100mV via 100k Ω

- See comments regarding specifications.
- Important to note that cross talk residual is clean, i.e., does not introduce distortion into the other channel----a common failing in many amplifiers and seldom mentioned.

Tone Controls

Bass ±10dB at 100Hz Treble ±10dB at 10kHz

Balance Control

+0.5dB to zero, each channel

Channel Matching 0.5dB

Filters

f. –3dB 50Hz	12dB/Octave
.f. – 3dB 5kHz	12dB/Octave

Subsonic, disc only, non-switchable (IEC 65) - 3dB 20Hz -8dB 10Hz -17dB 5Hz

Ultrasonic, all inputs except tape, non-switchable -3dB 60kHz -7dB 100kHz

e

Before making any measurements, the amplifier was pre-conditioned for 1 hour at 30% full power with both channels driven. Measurements were made after a further 5 minutes at full power.

Unless otherwise stated, the volume control was at maximum, tone controls at centre and filters switched out

The stability tests were made using various combinations of loads with capacitors up to $2\mu F$ and also with capacitors up to 2µF without additional load. Square wave signals were used over a frequency range of 20Hz to 20kHz.

The figures given are those obtained on the prototype amplifier.

Test Equipment

Hewlett Packard HP3580 spectrum analyser; Kron-Hite 4100 low distortion oscillator; Sound Technology 1700B distortion analyser, power output meter and low distortion oscillator; Teleguipment D83 oscilloscope; Polaroid oscilloscope camera.

This equipment was kindly made available by Armstrong Audio Ltd., at their research laboratory.

General Comment

Be very careful when comparing specifications of other amplifiers with the Winton. Many other published specifications use a weighted signal to noise figure. This would allow an extra 10dB or so to be added to their figures and tends to favour the poorer quality amplifiers at the expense of the really good amplifiers.

means that the circuit can be kept simple as no protection circuits are required, eliminating another source of distortion.

The Bi-f.e.t. Op.amps

The use of the Texas bi-f.e.t.s in the control unit section offers a number of advantages over the more usual 741 type of op.amp. The main ones being:

- 1) Wider bandwidth. 3MHz typical.
- 2) High slew rate $13V/\mu s$.
- 3) Low distortion 0.01%.
- 4) High input impedance j.f.e.t. input stage.
- 5) Low noise 18nV/Hz (TL072CP).
- 6) 80dB supply ripple rejection.

When used in low or unity gain circuits using large amounts of negative feedback the noise and distortion from the device is almost unmeasurable. The Winton amplifier has a basic sensitivity at the AUX inputs of 100mV and this has to be increased to 4V before the distortion reaches 0.1%. At any normally used input level the distortion is below the noise and completely inaudible.

Circuit Description

Both channels are identical and are pre-fixed "a" for left and "b" for right. Only the "a" channel will be described. See Fig. 1.

The Disc Pre-Amplifier

The input from a magnetic cartridge is fed into the base of Tr1, a low noise BC414 via the input load R1, $47k\Omega$, and an r.f. filter R2 C1. The transistor Tr1 forms one half of a differential input pair. Operating at a low current, approximately $100\mu A$ each half, ensures the minimum amount of noise.

The output from Tr1 is coupled directly to Tr3 a BC556. This in turn is coupled to a subsonic filter (as recommended by IEC65) consisting of C12, R16 and R17. This subsonic filter has its -3dB point at 20Hz, Fig. 2. This response, coupled with the normal RIAA response provides a 12dB/octave filter at subsonic frequencies and prevents intermodulation at low frequencies caused by warped records, etc., from being produced and affecting the overall sound quality. The RIAA negative feedback equalising circuit consists of R7, R8, C5, C7 and is connected between the output of Tr3 and the base of Tr2.

In order to ensure further the low noise factor in the disc amplifier, electronic decoupling is used, Tr13 (BC546) Tr14 (BC556) in each supply rail to remove any ripple or noise from the power supply. This supply has already been decoupled and stabilised by the Zener diodes, D1, D2.

The output from the disc pre-amplifier then goes to the selector switches S 1-5.

Control and Filter Section

From the selector switches S 1-5, each channel goes to one half of IC1 a bi-f.e.t. op.amp, Type TL072CP. This is used as a low noise buffer amplifier (with 4dB of gain) to provide a high impedance input, and a low impedance output suitable for driving the next stage, which also uses a bi-f.e.t. op.amp, Type TL082CP. This is used as an active high- (C15, C16, R22, R23) and low- (C17, C18, R24, R25) pass filter. Operating at unity gain, the filter is designed to provide a 12dB/octave cut-off with the -3dBpoints at 50Hz and 5kHz respectively, Fig 3. A slope of 12dB/octave is considered to be optimum for Hi-Fi use as a steeper slope could introduce ringing on transients, which would sound most objectionable. The tone controls follow the filters, again using a bif.e.t. op.amp, Type TL082CP, and the circuit is a Baxandall negative feedback type.

This type of tone control circuit keeps distortion and noise to a much lower level than the passive type and also provides frequency response contours more acceptable to the ear, Fig. 4.

The amount of bass and treble boost and cut has been restricted to around + or -10dB at 100Hz and 10kHz. Although it is possible to obtain up to 20dB boost and cut at these frequencies by removing R29 and reducing R33, extreme amounts of boost or cut are considered by the author to be bad design, as in practice large amounts can rarely be used. For example, suppose we have a signal requiring +10dB at 100Hz, with our circuit we can obtain this with only an extra dB or so at lower frequencies. If the unrestricted circuit was used, 10dB of boost at 100Hz would also produce 20dB of boost at 20Hz, as the response would continue to rise as shown in Fig. 4. The effect of this would be to increase all the rumble and other low frequencies a further 10dB over the required level at 100Hz. This would be in effect asking the amplifier for a considerable increase of power at low frequencies and as it cannot provide this, severe distortion would result. Note:

10dB of boost equals a voltage ratio of 3.16:1.

If our amplifier is already producing 50W into 8Ω , i.e., 20V, then $20V \times 3.16 = 63.2V$, power = $V^2 \div R = 3994 \div 8 = 499.28$ watts!!

Even with the amount of boost restricted to + 10dB, if the bass control is at maximum and the amplifier is just reaching maximum power (50W) at say 20Hz then the maximum power at 1kHz must be limited to 5W (assuming a flat frequency response of the input signal). In practice, the tone controls are normally used to make up for the deficiency in the incoming signals, i.e., to restore the signal to an overall "flat" response, so the power restriction normally won't apply.

The output from the tone control circuit then passes to the balance control and the tape monitor switch S8. This switch selects either the output from the control unit section or the output from a tape recorder (connected to the tape socket) and feeds the signal via the volume control to the power amplifier. A tape output signal is permanently connected from the control unit output to both the tape socket and via $100k\Omega$ resistors to the AUX 1 socket. This arrangement allows the use of either or both reel to reel and cassette recorders with the Winton. When using a reel to reel or cassette recorder fitted with a monitor head (plugged into the "tape" socket) tape monitoring can be achieved by simply pressing the "tape" button.



Fig. 2: Disc input — Subsonic filter response



Power Amplifier

The circuit used for the power amplifier is based on a design produced by Hitachi for their power m.o.s.f.e.t.s. The original circuit has been modified by the author to suit the requirements of the Winton and electronic decoupling added to further improve the overall specification.

The signal from the volume control, VR4 is fed via an r.f. filter R40, C27 to the base of Tr4. This is one half of a differential pair Tr4, Tr5 using low noise transistors Type BC556. These in turn drive a second differential pair, Tr6, Tr7, Type 2SC1775, which has an active collector load (current mirror formed by Tr8 and D3), to maintain the pushpull action. The power m.o.s.f.e.t.s Tr11, Tr12 are driven directly from the second differential pair. As the

inherent distortion of the amplifier is very low, only 45dB of negative feedback is used, this is fed from the output, back to the base of Tr5 via R46, R45, C31. The use of less feedback means that, at the overload point, the onset of distortion is less severe then would be the case with amplifiers using considerably more negative feedback. Diodes D4, D7 and Zeners D5, D6 form an overdrive protection circuit to prevent the gates of the m.o.s.f.e.t.s from receiving excessive drive voltages in the event of a fault condition.

Due to the excellent high frequency response of the amplifier only minimal feedback phase correction is required via C31, R45, and L1, R58, R59, C39. The amplifier is unconditionally stable.

As mentioned earlier, electronic decoupling has been included in each power amplifier to reduce the power

* components

Resistors	Capacitors
	4 2·2μF 03V - 10 - 013, 14, 20, 23, 20
2200	47 ± 637 10 C2 3 8 10.33 34 44 45
3300	100µF-16V 6 C4:28 29
560Ω 4 855,56	220µF16V 2 C30
820Ω 2 . R6	4700µF 63V 2 C40, 41 (Note: 7A ripple rating)
1kΩ 6 R2, 10, 20	
2·2kΩ 4 R9,40	
- 2.7kΩ 6 R32, 42, 43	Polystyrene
3.3kΩ 2 3 R33	10pF 631
5 6LO 1848 1 5 5 1 54, 5	
8.2k0	320-E 4 - C17 18
10k0 2 B11	330dF 2 C6)
12kΩ 2 / R B46	3 3nF 4 C5.23 2.5%
15kΩ 2 R45	• • • 10nF 2 C7)
18kΩ 2 R48	
22kΩ 2 , R8	
33kΩ 4 R22, 24	Polyester
39K12 2 F141	
47KS2 10 H1.3;21,29,34	
100k0	a 47nF 4 C21 22 (Note: matched 2~5%)
$120k\Omega$ 4 851 53	10nF 2 C9.11
$150k\Omega$ 2 R18	
390kΩ	
	Semiconductors
	Diodes
5%-0-5W	Til209 15 D8
10Ω 2 R59	BZY88C15 +6 U1, 2, 5, 6
IUKS2	
	NDL-UZ
5% 1W	and the second
10Ω 2 858	Wansistors
4.7Ω 2 RB1	8C414 4 Tr1, 2
1kΩ 2 R37, 38	BC556B 11 Tr3, 4, 5, 8, 10, 14
	BC546B 3 Tr9, 13
Bian	2SC1775E 4 1r6, 7
330 7 060	200130 4 HTT 20148 2 Tr12
	Integrated C' cuits
Potentiometers	TL072CP 1 IC1
100kΩ + 100kΩ lin 2 VH1.2	TLD82CP 2. IC2, 3
100kΩ + 100kΩ log 3 1 VH4	
100kΩlin, 1 VR3	
	Wiscellaneous
Min Preset Borizontal mounting	Belclere 6854/3 (T& T Electronics). Chassis and front
1k0 4 VR5.6	oanel, Winton (T & Electronics), Case Winton (T & T
	Electronics) Pins, RS 433-630 (47), Fuse holder, RS
	412-879 (3); Fuse 2-5A Quick Blow (2); Fuse; 1A
	Slow-Blow (1); Pillars, RS 543-737 (5); Cable Ties,
Switches	RS 543-349 (8), Cable Ties, RS 543-428 (6); 4mm
Win, toggle d.p.d.t. 2. 59, 10	Ferminals, Black (1), Blue (2), Red (2), DIN Socket
nush putton unit o keys 1 - 53 10 01	The recomment of way rout 10/, Jduk Suukel no 470-012(1).
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Internal view of the *PW* "Winton" amplifier. Here, we show the final version using a printed circuit board. The model illustrated on our front cover is the prototype, which was constructed on perforated board with hard wiring

supply noise and ripple voltage, as well as improving the cross-talk (Fig. 5) between channels. Transistors Tr9, BC546 and Tr10, BC556 are used for the decoupling circuit.

Provision is made to adjust the d.c. offset at the output to zero. By adjusting VR5, the current through the differential input stage can be adjusted to compensate for the slight gain variation in each half of the differential pairs. It is very important that the output from the amplifier does not contain any d.c. voltage, as this would cause the loudspeaker cone to take up a position away from its centre which in turn would cause distortion because of the resulting non-linear cone movement.

Power Supply

By no means least important, the power supply of the Winton has been subjected to the same standard of design as the rest of the circuits, although on the surface it may look simple.

In a high fidelity amplifier the power supply can either turn a good amplifier into a top quality model or into a not so good one. There are two main reasons for this:

- 1) The regulation of the supply volts must enable the amplifier to reach its full power potential under the worst possible signal conditions. At the same time, it must not let the voltage rise to dangerous levels under quiescent conditions.
- 2) The external magnetic field from the transformer must not inject any hum into the amplifier circuits, particularly the disc amplifier.

To enable the full performance that the Winton is capable of to be achieved, it was decided to use a toroidal type of transformer. The design and development of this special low-field transformer was carried out by Belclere Ltd., and this transformer is now available from T & T Electronics.

The output from the transformer goes to a bridge rectifier and then to the two 4700μ F capacitors C40, C41. These, and the electronic decoupling circuits used in the

power amplifiers, coupled with the Zener stabilised supply to the control unit and further electronic decoupling to the disc pre-amplifier, ensure that only about 0.001% of the total output from the Winton is noise from the power supply. Even on the disc input the total noise in the output from all sources combined is only about 0.04%.

Heat Sink Ratings

When designing an amplifier for domestic use, the design is almost always a compromise between all the various requirements. In the case of the Winton amplifier this applies to the size of the heat sink and mains transformer.

It is possible to make both of these components in such a way that the amplifier could deliver 50 + 50Wcontinuous power for hours on end without much temperature rise. The cost of doing this would be so high as to put the amplifier out of reach of most peoples' budget. On the other hand these components could be made down to a low price and the amplifier allowed to run very hot, which would lower the reliability.

In the design of the Winton the author has arranged that these two components will allow the power f.e.t.s and the transformer to run well inside their respective maximum rating under continuous drive conditions, although under these conditions both the heat sink and transformer will reach quite high temperatures and certainly be too hot to touch.

When used for its normal purpose, i.e., reproducing music, the temperature rise is very much less and after some hours of loud music will still be only moderately "Hot". The actual heat sink may reach 60°C under these conditions.

This compromise doesn't reduce the quality of the reproduced sound in any way, but it does avoid a severe pain in the wallet.

NEXT MONTH Constructional details



Those who built the Avon transmitter described in PW, July 1978, may be interested in one or two simple yet significant modifications which were ultimately incorporated into the prototype. The purpose was to increase the power output to approximately 20 watts for the original 24 volt supply and to give some consideration to a repeater access facility. The question of powering the transmitter from 12 volts—a car battery, for instance—was also examined, and the update caters for this at a reduced r.f. output of around 9 watts.

In the first place, the "Calibrate" facility provided by S1 in the switching arrangements of Fig. 10, p. 52 (*PW*, Aug. 78) was dispensed with and the switch re-assigned to a toneburst facility for repeater working. Details of the toneburst module appear elsewhere in this issue: the board on which it is constructed can easily be accommodated in the available space.

Step two is to replace the 15V regulator (7815) with a 12V type 7812. Our attentions are then directed towards the power amplifier board, where the majority of changes take place. Here a little care is needed, although the modifications are far from complex. The revised board could, of course, be constructed as an "afterburner" which would provide a substantial increase in output from, say, a hand-held tranceiver.

Initially, remove the resistor R1 (82 Ω 2W) and take the h.t. feed (now 12V) directly to the board. The collector supply to the p.a. transistors remains routed to the *input* of the 7812 voltage regulator, via the switching of Fig. 10 (Aug. '78).

Now take out the BLY83 transistors and carefully expand the holes in the p.c.b. to accommodate the new 2N5642 devices. The leads will require a little trimming and this should be done fairly accurately to permit the transistors to fit on the appropriate islands on the board.

The increased output of the new p.a. inevitably means that appreciably more heat will be dissipated by the final stages. Additional heatsinking will have to be provided and in the prototype this took the form of a piece of aluminium $100 \times 50 \times 4$ mm, drilled to accept the studs of the power transistors. The sink is fixed to the main chassis by means of a suitable nut and bolt in each corner and the surface in contact with the chassis coated with Thermopath grease or similar compound. Insert the 2N5642 devices and tighten down, carefully soldering the connections to the p.c.b. afterwards (Fig. 1).



Fig. 1: Mounting the power transistors

The installation of the toneburst module designed by Philip Hodson is extremely simple. A power supply is taken, via the defeat switch S1, from across the energising coil of the keying relay RLA, which is shunted with a diode to absorb switching transients. Miniature screened lead feeds the signal input to the base of Tr2 (Board 1), with a 1 μ F tantalum capacitor placed in series.

The prototype exhibited some problems which ultimately were attributed to stray r.f. entering the microphone input. It is therefore recommended that a



Fig. 2: The new switching arrangements — note the earth on changeover contacts RLA2, the option of different coil resistances for RLA and RLB from the original article, and the addition of diodes D2 and D3

continued on page 53



A wideband r.f. noise generator is a most useful device for carrying out rapid performance checks on all types of h.f. and v.h.f. receiver. The wide variety of applications of such an instrument includes gain checks, receiver sensitivity measurements and, where a calibrated attenuator is available, the noise generator may also be used to carry out accurate measurements of gain, noise figure, a.g.c. characteristics, and the calibration of signal strength meters.

The noise generator described here uses a minimum of components, is simple to construct and, although its output level is uncalibrated, is eminently suitable as a source for receiver alignment with an output which is substantially of constant level over a very wide frequency range. A previous article dealt with the construction of a matching calibrated 50Ω attenuator.

True "white" noise can be thought of as a signal which is evenly distributed over an infinitely wide range of frequencies and, although the instantaneous voltage varies randomly, the r.m.s. noise voltage developed into a resistive load will be constant when measured over a short time period. The noise power developed into a purely resistive load is thus directly proportional to the bandwidth in which it is measured.

Generating Noise

The simplest form of noise source is that due to thermal agitation current in a conventional carbon resistor. The fluctuation in current caused by the random movement of electrons is evenly distributed in frequency but the noise voltage produced is usually extremely small. Practical noise generators do not provide an infinite noise spectrum. This is due both to the limitations of the noise source itself and to the bandwidth of any following amplifier. The choice of noise source depends on a number of factors.







Thermionic diodes are popular in commercial noise generators but, for economy of supply and portability, the noise source employed in this design is a conventional Zener diode. A considerable noise voltage is developed across the series load resistor of a Zener diode when it is undergoing breakdown. It is, however, usually necessary to include a stage of amplification following a noise source so that its output voltage is of a sufficient level for general purpose use. This has the added advantage that the amplifier stage also provides a degree of isolation between the noise source and the input of the circuit or receiver under test.

Circuit Description

A block diagram of the noise generator, together with its Thevenin equivalent circuit, is shown in Fig. 1. The noise output of the Zener diode is amplified using a twostage wideband amplifier, Tr1 and Tr2. Both transistors operate in the common emitter mode and the gain of the second stage, Tr2, is made variable by the application of negative feedback by means of VR1. The device used in the Tr2 position has an effect on the bandwidth of the noise output. For use at v.h.f. the transistor recommended is a BFY90 as this has a very high cut-off frequency. For general purpose use at h.f. (i.e. to at least 30MHz) a much less expensive transistor may be used, and the 2N706 is quite adequate.

In order to facilitate matching to a standard 50Ω system, the output of the wideband amplifier is terminated in a 56 Ω carbon resistor. This, together with the parallel effect of the collector load resistor of Tr2, ensures an output resistance of almost exactly 50Ω . The output noise voltage can be thought of as being generated in this resistor and thus the Thevenin equivalent circuit of Fig. 1b is realised.



Fig. 2: Circuit diagram of r.f. noise source

Construction

Construction is straightforward and the use of the printed circuit board layout described is highly recommended. Where a printed circuit board is not used, care must be taken to ensure that all wiring is neat and as direct as possible. The noise generator is housed in a standard diecast box. Any alternative screened metal box can be used provided that the leads from the printed circuit board to the output socket are kept as short as possible. A small



Fig. 3: Internal arrangement of components

***** components

Resistors	k X	4	N.		ń	N.	ş.	s;	S.	ŝ.	S.
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hole is made in the front panel of the box to provide access for the adjustment of VR1. Either a skeleton type or a knob type pre-set may be used. The latter will prove to be more satisfactory where frequent adjustments of output level are likely to be made. Care should be taken to ensure that the hole made in the case aligns correctly with the adjusting slot or shaft of the pre-set. Standard variable resistors are not recommended for use since they are not only too large but have considerable stray reactance which may impair the frequency response of the amplifier. The BFY90 is available in both TO18 and TO72 packages the latter type should have pin 4 (case connection) clipped off before soldering the remaining connections to the board.

Testing

When the wiring of the noise generator has been completed, a thorough visual check should be made before connecting the battery. Connect the noise generator to the input of a receiver and check that an output is obtained. If the receiver has a signal strength meter this should show a steady indication; if no signal strength meter is fitted a 'rushing' sound should be heard from the loudspeaker. Adjustment of VR1 should cause a change in the noise level. If possible, a check should be made over a range of frequencies. If the receiver sensitivity is constant over a range of frequencies its signal strength reading (or audible output level with a.g.c. off) should remain constant as the receiver is tuned across its entire frequency coverage. Receiver alignment may be carried out using the receiver's signal strength meter in a conventional manner. The use of an a.f. voltmeter connected across the loudspeaker terminals as an output level indicator is strongly recommended when alignment is carried out on receivers not fitted with a signal strength meter.



Amongst the majority of would-be and newly-licensed amateurs there still seems to exist an air of uncertainty with regard to repeaters. This is quite apparent by the number of new 2m stations openly admitting "I'm recently licensed and don't fully understand repeater operation". This article looks at the principle of repeater working and suggests a suitable design for a toneburst generator with which to access them.

Operation via Repeaters

A repeater may be simply described as a remotelyoperated transceiver which simultaneously transmits on one frequency and receives on another. Such a function is usually referred to as *duplex*, as opposed to *simplex*, when transmission and reception occur on the same channel but not at the same time. In the duplex mode, the repeater transmit and receive frequencies are well separated, the transmit channel being the higher of the two. This difference is referred to as the "shift" and is 600kHz in the case of 2m and 1600kHz at 70cms.

The amateur therefore would normally transmit on the repeater receive channel, which is known as the input. Some confusion arises here, especially when the term "reverse repeater" is used. This, as the name implies, refers to the amateur transmitting on the repeater **transmit** channel, effectively by-passing the facility. The function can be useful when calling a station known to be within point-to-point distance, and under normal circumstances you would transfer to one of the simplex channels when contact has been established. For obvious reasons you should not call a station in this mode unless you can hear him on the repeater input.

Channel	Frequency (Receive)	Frequency (Transmit)
RO	145-000	145.600
R1	145-025	145-625
R2	145.050	145-650
R3	145-075	145-675
R4	145.100	145.700
R5	145-125	145.725
R6	145-150	145.750
B7	145.175	145.775
R8	145-200	145-800
R9	145-225	145-825

Table 1: 2m Repeater Channels

Practical Wireless, March 1979

Table 1 shows the frequency and channel allocations for duplex repeater operation. Not all of these are at present in use within the UK however, our own stations being between R3 and R7 inclusive.

Access to a repeater is gained by a short burst of tone (1750Hz) generated at the beginning of the transmission and lasting for 500ms typically. Some repeaters require this to be followed by several seconds of speech without loss of carrier, and this is quite difficult to arrange if all the switching is done manually. The device described in this article fulfils all these functions by one operation of the push-to-talk switch.

Most repeaters, with the notable exception of GB3LO (London), restrict the duration of transmission. After a pre-determined period a *time-out* occurs and the repeater releases, having first sent a series of "pips" to advise the receiving station of its intention.

When the input of the repeater senses the absence of a carrier, an advisory signal is sent, indicating that the station is ready to be re-accessed. Here it should be clearly understood that the terms "access" and "re-access" do not relate to the same function. Re-access involves the resetting of the timing mechanism immediately after the repeater has been used, whereas access is given to mean "starting from cold", as it were.

British repeaters vary considerably in the method by which re-access takes place. Some require a toneburst with or without speech to follow, others operate on receipt of speech alone whilst some require only to receive a carrier, whether modulated or not. Some very interesting information is given in the *International VHF-FM Guide*⁺, for those who would like to consider individual cases.

The Toneburst Module

Referring to the circuit of Fig. 1, we can see that the crystal oscillator operates at a frequency of 910kHz. The Schmitt trigger IC2a converts the waveform into a sharpedged square wave which is fed to the twelve-stage ripple counter IC3 as the clock reference. This counter, in conjunction with other gates, divides the clock frequency by 520, thereby obtaining the required 1750Hz necessary. The output is routed via another Schmitt trigger IC2c into a 3-pole Butterworth filter, which produces a relatively pure sine wave.

+ Available from J. Baldwin (G3UHK), 41 Castle Drive, Maidenhead, Berkshire (90p post free).



Immediately power is applied, the capacitor C9 starts to charge through R10. As the potential rises towards that of the supply, a point is reached where IC2c changes state, effectively blocking the path of the 1750Hz signal. When this occurs, the system has to be re-set by disconnecting the power supply and consequently allowing the capacitor to discharge.

By installing the toneburst module in such a way that on pressing the microphone p.t.t. switch power is simultaneously applied to the transmitter, the access signal will be generated on the initiation of each "over" and the possibility of dropping carrier is eliminated. The provision of a switch in the power supply to the module permits its isolation when working simplex.

★ components

Construction

The discrete components should be mounted initially, leaving the crystal and integrated circuits until last. Omit resistor R12 at this stage, otherwise it will have to be removed in order to test the board.

If you do not wish to use holders, leave IC 1–3 until last. These devices are c.m.o.s. and so should be handled as infrequently as possible prior to installation. The techniques for soldering these components have been considered on many occasions within these columns and in other publications. These should be closely adhered to if damage is to be avoided from static charges. The crystal is also susceptible to damage from excessive heat. The best method of soldering this is to use a very hot iron and to complete the process as quickly as possible.

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Fig. 2 (left): Copper track layout of the p.c.b. shown full size and (right) the component layout

The capacitor C9 is specially selected, and it has been found that only the type in the components list is suitable. Most tantalum beads have a tolerance of 20%, and these gave rise to problems in the prototypes. For this reason a tubular version with axial leads and a tolerance of 10% is required. Increasing the value of this device will lengthen the duration of the toneburst.

Testing the Board

Make certain that you have not fitted R12 before commencing. Now connect a small loudspeaker or pair of headphones between the slider of VR1 and earth, then temporarily short across C9 with a small piece of wire. Apply around 15 volts and you should hear a continuous 1750Hz tone. Disconnect, and remove the short across C9. Re-apply power, and the tone should occur for a period of approximately half a second. Should it happen that the burst is appreciably longer, then reduce the value of the capacitor slightly. Selection of the correct value proved to be the only variable in the prototypes. The final decision to use 4.7μ F was taken only after exhaustive experimentation, which indicated that this value produced the most consistent results. Provided the burst duration falls between 540 and 650ms, no difficulties should be encountered.

After carrying out these procedures, don't forget to insert R12!

If you are installing the module into the Avon transmitter, the following information may be helpful. Connect a microphone to the transmitter, key the device and adjust VR1 on Avon Board One to obtain a clear and undistorted speech output. Do not attempt to set for maximum volume; this is not the function of the deviation control.



Fig. 3(a): The toneburst module fitted to a simple push-to-talk arrangement



Fig. 3(b): Fitting the module into a high-power transmitter, where a keying relay is used

Too little deviation is preferable to an excess. Once adjusted, do not touch this control again, but alter VR1 on the *toneburst* module to give a level of tone slightly lower than that obtained with speech. If you are lucky enough to have access to an oscilloscope, the speech should be set for a maximum deviation of 5kHz and the toneburst limited to 3.5kHz.

Follow-up to the PW Avon

continued from page 48

100pF silver mica capacitor be connected from the input to earth, as close to the p.c.b. as possible, to provide the necessary decoupling.

Replace the p.a. board and the modifications are now complete. Retuned, the transmitter will produce about 9 watts of r.f. with a power supply of 12 volts, rising to around 20 watts for the original 24 volt supply.

Operating the microphone pressel switch with the "Access Repeater" facility enabled will automatically generate toneburst, thus providing the required conditions to open repeater stations.

The power supply featured in the September '78 issue of PW can easily be modified to handle the increased current drain, provided the transformer is capable of delivering it. Merely place another 0.47 Ω 25 watt precision wirewound resistor in parallel with R2 (Fig. 1, p. 30, September '78 issue).

SY TO	IN THIS	ST KIT ISSUE
2.16	4093	46p
27p	4040	65p
24p		
	2.16 27p 24p	2.16 4093 27p 4040 24p



When using a small current (such as that from a t.t.l. or c.m.o.s. logic device) to control a much larger current in a load, it is possible to employ a simple transistor amplifier in some cases, whereas if a higher current gain is required a power Darlington device may be used as an amplifier. However, simple circuits employing such devices do not incorporate any means of limiting the current in the load so as to protect both the load and the amplifier device from possible damage. Also they do not incorporate any means of protecting the amplifier device if it should undergo an excessive temperature rise.

The TDE 1607 is a new integrated circuit from Thomson-CSF which can be used to overcome these problems. Basically it is a protected operational amplifier, being designed for fairly high currents and voltages, and is specifically intended for the control of the current in such loads as relays, lamps and stepping motors.

The TDE 1607 is essentially fail-proof in operation provided that the absolute maximum permissible supply voltage and input voltage (both 36V) are not exceeded. An external resistor can be incorporated in the circuit so as to limit the current in the load to any chosen value up to 0.5A (although the TDE 1607 has a maximum permissible output current of 1A). If the silicon chip in the device becomes too hot for safety, the internal circuit will automatically reduce the output current and thus prevent possible damage.

Connections

The TDE 1607 is supplied in a small circular metal transistor-type package with 6 leads. The connections are



Fig. 1: Package outline and connections



Fig. 2: Block diagram and "pass" transistors

shown in Fig. 1, but it must be remembered that this is a **top** view. This device can operate over the wide temperature range of -25° C to $+85^{\circ}$ C.

The way in which the device operates can be understood using the block diagram of Fig. 2. When the non-inverting input to the amplifier (marked +) is at a voltage greater than that of the inverting input at pin 2 (marked -), the output voltage from the amplifier at point A in Fig. 2 will be relatively high (only a little below that of the V⁺ supply line).

The output from the amplifier will then normally feed a current into the base of the transistor Tr2 (inside the TDE 1607) so that this transistor conducts and passes a current through the external resistor R_{sc} and through the load, R_{L} . The value of the current sensing resistor R_{sc} determines the maximum current which can flow in the output circuit. For example, if the value of R_{sc} is about 1.5Ω , a current of 0.5A will produce about 0.75V across R_{sc} and this voltage is applied between the base and emitter of the transistor Tr1. It is adequate to cause this transistor to commence to conduct, so instead of the current from A passing to the base of Tr2, it will pass to the collector of Tr1. Thus Tr2 will pass only just enough current to maintain a large enough voltage across R_{sc} to keep Tr1 conducting. The current passing through R_{sc} is equal to the load current, so the internal transistors Tr1 and Tr2 can be used to limit the load current.

If R_{sc} has the value of 1.5 ohms mentioned above, the load current will be limited to about 0.5A, whereas as R_{sc} is increased to 3 ohms, the maximum load current will be

about 0.25A. The writer found experimentally that still larger values of R_{sc} could be used, a value of 100 Ω producing a maximum current of some 10mA. The exact value of the maximum current varies somewhat with the case temperature of the device, typical values being plotted in Fig. 3. Obviously the output current will never exceed the value of V⁺ divided by ($R_L + R_{sc}$), since an adequate voltage is required to cause the current to flow.

The output current will flow only when the noninverting input of the internal amplifier (pin 3) has a potential exceeding that of the inverting input (pin 2); if this condition is not satisfied, the device will be 'off' and little output current will flow. If the silicon chip of the device ever becomes too hot for safety, the thermal protection circuit of Fig. 2 passes current from point A to ground so that there is little bias current to feed the internal transistor Tr2; the output current therefore falls until the device cools. However, it is unwise to operate the device for an appreciable time at such a high temperature that the thermal protection circuit is in operation, since such high temperature operation can tend to impair device reliability and certainly tends to produce surface defects on the chip. It should rather be regarded as a safety circuit which is not normally used.



Fig. 3: Typical output current plotted against case temperature

Power Supply

It is recommended that TDE 1607 device is fed from a power supply of between 10V and 30V; it is, of course, possible to use $\pm 15V$ balanced supplies if desired. The supply current is typically only 3mA (maximum 5mA for any TDE 1607 device) from a 24V supply, when the output current is zero.



Fig. 4: Graph showing safe operating area

When the output voltage from the device is low, ideally no output current should flow and the output voltage should be zero. In practice, an output leakage current of up to 100 μ A can flow through the load at a junction temperature of 20°C or up to 500 μ A at a temperature of 85°C, whilst the output voltage may be a volt or so above the potential of the ground line. Similarly, when the output voltage is in the high state, there may be a potential of up to 1.8V between the output and the positive supply line.

The input currents required by the TDE 1067 are quite small, typically 100nA (maximum 1.5μ A). The input offset voltage at which switching from one state to the other occurs is typically 2mV (maximum 50mV) between the two inputs.



AD262 *D1: Required if inductive load used



In order to provide adequate protection for the device, the value of the resistor R_{sc} should be chosen so that one operates within the unshaded area of the graph of Fig. 4. For example, with supply volts of up to 24V the value of R_{sc} should not be less than 0.8Ω so that the output current is never allowed to exceed about 1A. Obviously there is no objection to the use of larger values of R_{sc} where a smaller output current is adequate for the application concerned.

Typical Circuits

A typical application circuit for the TDE 1067 is shown in Fig. 5. The inverting input is biased by the potential divider to the desired value; a load current flows when the potential of pin 3 exceeds that of pin 2.

The diode D1 is required only if the load is inductive, such as a relay. When the current ceases to flow through the relay or other inductive load, a short transient reverse voltage can be developed across the relay coil. The diode is used to short this transient voltage to ground, since transients can damage the TDE 1067 device.



Fig. 6: Extension of the Fig. 5 circuit providing higher output current

5A Output

To provide a higher current than the TDE 1067 can itself supply, the circuit of Fig. 6 may be used in which an extra *pnp* power transistor fitted to a heat sink is used with the TDE 1067. Current limiting is again provided by the resistor R_{sc} , but the value of this component is much smaller owing to the higher current passing through it. Such small resistors are conveniently made from a length of resistance wire wound around a high wattage resistor of much higher value.

Conclusion

The TDE 1067 is a versatile device which can be operated by very low currents to its inputs. When used correctly in circuits, it is almost impossible to damage the device even if the output is short-circuited to ground. At the time of writing the price is $\pounds 2.99$ (including VAT) plus $\pounds 0.20$ for packing and postage from Phoenix Electronics Ltd, 46 Osborne Road, Southsea, Hants PO5 3LT. KINDLY NOTE!

Car Radio LW Converter, December 1978. The MVAM115 varicap diodes are shown on the circuit diagram but omitted from the components list. They are available from Ambit International, Gresham Rd., Brentwood, Essex.

Breadboards Supplement, December 1978. On the 5V regulated supply (p. 2 of supplement), D1 is shown incorrectly. The configuration should be identical to that of D2.

Digital Door Chimes, December 1978,

In the audio amplifier section in Fig. 1, Tr3 (BC160) is shown inverted. The emitters of Tr2 and Tr3 should be linked. The collector of Tr3 goes to chassis.

Sandbanks Metal Detector, January 1979.

C17, 18, 19, and 20 are incorrectly shown in the components list (p. 50) as 10nF. These should all be 1nF, as correctly shown on the circuit diagram on p. 49. In the circuit diagram, Fig. 2, diode 3 should be reversed in polarity.

PW "Purbeck" Oscilloscope

Some readers seem to have misunderstood the information given in the "Follow-up" article (p. 55 January 1979) regarding the mains transformer.

Due to an unfortunate combination of circumstances, the differing characteristics of the transformers supplied by Watford Electronics with their PW "Purbeck" kits were not picked up by the author when incorporating the sample transformer in the original PW "Purbeck" oscilloscope. This oscilloscope and transformer have been in daily use now for many months, but two changes have proved necessary as follows.

If you are using one of the original Watford Electronics transformers then, as stated in the follow-up article, all that need be done is to replace R103 by a wire shorting link, and to reduce the value of R101 to around 120Ω (5W wire-wound); or better still, replace R101 by a wire shorting link. This provides a greater margin for low mains voltage. It is NOT necessary to change the transformer.

Don't forget to check, as stated previously, that the X and Y Boards draw 20mA and 30mA respectively from the $\pm 150V$ stabilised supply.

Please note that, when adjusting the input attenuator trimmers (C3-C10), a metal-bladed screwdriver is not suitable. An insulated coil-core adjuster should be used, and the hand held well clear to reduce stray capacitance effects.

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(October 1978 PW)

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Wire-wrapping

OK are probably the only company who produce wire-wrapping equipment specifically for the hobby market and their latest kit HW-KI brings power wire-wrapping within economic reach of the home electronics enthusiast.

The main item in the kit is a newly designed battery-operated wire-wrapping gun, based on the design of the company's industrial units. It is for use with 0.6×0.6 mm mini-wrap terminals and has a bit and sleeve to give modified wrap. The tool is also self indexing and has a back force device to prevent overwrapping.

It is powered by two NiCad batteries, the two batteries provided having a year's guarantee, and a mainsoperated charger is included. Also in the kit is a handy 'pocRet-sized' wire dispenser, containing 50ft of 0.25mm

from the copper cladding around the hole. The application is to provide clearance spots for ground-plane mounting of d.i.p. sockets, feedthrough sockets, etc.

In each case the tools comprise a replaceable high speed or carbide twist drill, either 0.0400in diameter or 0.0292in diameter. Each tool is also available in three pad milling sizes, 0.01in, 0.15in and 0.20in. Prices and a free catalogue are available from: *Carel Components, 40–44 The Broadway, London SW19 1SQ. Tel:* 01-540 7186.

Mini-drill

Recently introduced by Boss Industrial Mountings Ltd. is a mains drill plus accessory kit housed in a specially designed carrying/presentation case with transparent lid.

This small but powerful 220/240V a.c. BIMDRILL is supplied with 4 collets capable of accepting tools with shanks of 1mm, 2mm, 2.4mm and 0.125in diameter, and will readily drill brass, steel, aluminium and p.c.b.s.

Complete with spring-loaded on/off switch and a 2 metre long cable fitted with a 2-pin DIN plug, this BIMDRILL has a fully insulated, high impact, ABS body, weighs less than 250g and has an off load speed of approximately 7500 r.p.m.

The accessories include 20 assorted twist drills, mops, burrs, grinding wheels and mounted points, all individually and securely located within the $230 \times 130 \times 50$ mm case. Priced at £22.14 which includes VAT and P&P, the kit is available from: *Boss Industrial Mouldings Ltd., Higgs Industrial Estate, 2 Herne Hill Road, London SE24 OAU. Tel: 01-737 2383.*

wire. The other item is a hand tool which can strip wire, wrap and unwrap.

Suggested retail price of the HW-KI is £46·28 (inc. VAT and carriage). *OK Machine & Tool (UK) Ltd., 48a The Avenue, Southampton, Hants SO1 2SY. Tel: (0703) 38966/7.*





Practical Wireless, March 1979

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by Eric Dowdeswell G4AR

An interesting letter from Les Light G3KDL of Wembley, Middx, prompts me to write a few words for the reader who may be reading this column for the first time, perhaps after having had interest aroused elsewhere by contact with amateur radio.

Les received a letter from a young but obviously enthusiastic listener simply saying he had heard Les on the 80m band. Time and frequency were given, but little else, except some details of the receiver and aerial. Now, with the best will in the world, the report was not of much use at all, but Les, no doubt thinking back to his own early days in amateur radio, did send a QSL.

It is naturally very exciting for the newcomer to listen to amateurs for the first time especially when they realise the comparatively low power of the amateur, possibly being accustomed to the high power stuff on the BC bands. So off goes a report such as quoted above, with a request for a QSL, and probably no s.a.e!

With the advent of so much commercial gear and highgain beam aerials, worldwide working by amateurs is commonplace and the excitement of working and then getting a QSL from some DX spot has waned somewhat. QSLing had decreased over the years to the extent that some amateurs don't even bother to have cards printed!

So the report from a listener, even if from the other side of the world, is of little use, generally speaking. It has got to be something unusual in the reception, or perhaps the other fellow is using very low power, before it is worth sending a report. I would counsel any beginner to first get the *Guide to Amateur Radio* from the RSGB to see what it is all about and to study, in particular, the section dealing with QSL matters.

The cost of QSL cards and postage today is enough to deter anyone from the hobby, so if you must send a report and request for a QSL, then make it easy for the other station by sending a s.a.e., or International Reply Coupons for overseas stations, and a report worthy of attention. Don't use these silly 5×3 in envelopes! Send something a bit bigger. Having sent tens of thousands of QSLs out when I was ST2AR I rate the small envelope as the biggest annoyance of all!

Collecting QSLs can be fun, or used to be, but if there is any intention of going on to get a transmitting licence I would say the best policy is to spend the money on the receiver and associated gear rather than on QSLs and QSLing. You'll need every penny for the entrance fee for the RAE and the ticket itself. Then you can have your own QSL card, and receive them from stations you work, when they really mean something.

General Notes

An interesting note from **Owen Frame** G4EIF of Reading, Berks, on the mysterious H44LW mentioned in the December issue. Owen confirms that it is the Solomon Islands, as he has a QSL for an s.s.b. QSO. QTH is Box 19, Honiara. Owen added, that he used to be ST2WF in Khartoum in 1934 with the RAF, with 80W of c.w. from 12V accumulators! As he says: "happy days".

N. Eddy writes from Truro, Cornwall, for the first time although he has been reading PW for many years. He is a member of the Cornish ARC and sports an AR88 and BC348 with main interest on 14MHz. He also has a 62H receiver which he'd like to convert for 2m operation, if any reader can help him on mods. Write to: Little Tregadles, Laity Moor, Ponsanooth, Truro, Cornwall, all expenses paid of course. John Bell BRS40279 in Melksham, Wilts, had "quite an amazing response" to his request for info on the AR88, particularly from Michael Swain G8MMP. John recently got hold of an Eddystone EC10 and, as an instructor to mentally handicapped people, he is going to try to teach elementary geography using the EC10 to demonstrate the reception of BC stations from various parts of the world. Apparently a fellow instructor has had some success this way.

Newcomer to the column **Derek Brabrook** of Laurgharne, Dyfed, Wales, has been using a domestic type Pye set for a couple of years or so, as it goes up to 26MHz. However, he finds his 223ft aerial "difficult and poor". I'm not surprised! It must be overloading the front end I imagine on the stronger signals, as I doubt whether there is any r.f. gain control, and selectivity can't be much good on such an old set. Instead of blaming the set, I suggest keeping the aerial and getting a more modern set! In Oswestry, Salop, **David Wyatt** aged 14, has acquired

In Oswestry, Salop, **David Wyatt** aged 14, has acquired a BC348 and promptly found KC4USX in Antarctica on 20m s.s.b. His main regret is that it does not go above 18MHz so he can't listen to the 10 and 15m bands. Again, a case for a converter. David offers to reply to any readers who'd like to write to him on the BC348 at 11 Prince Charles Road, Oswestry. You might be overwhelmed, as the receiver was a very popular one OM! David is also learning the code from records and threatens to send in c.w. reports before long. We are certainly short of them of late; hint, hint, to other readers!

Round the Bands

Very little in the way of logs this month, unless the season's rush on the PO has held them up in the pipeline. Could be everyone's waiting to see if Father Christmas has remembered them with a nice, shiny, new receiver! **Bernard Hughes** BRS25901, who hasn't written in for a while from Worcester, liked my article on receiver accessories and wants more! Drop a line to the Editor OM! In spite of a new 20m dipole Bernard found the most interesting DX on other bands with his Drake receiver. Latest QSL received of any note was from PYORO on St Peter & Paul Rocks. Worthy of note in his log were HS1ABD, STORK (!) VR3AH and XT2AT on the 10m band, KJ6BZ VP2DAY on 15m and HR1HMV, YS1RRD and 4W1BC on 40m, all s.s.b.

From Leigh-on-Sea, Essex, Ian Marquis A9140 keeps up the good work keeping his ears open on all the h.f. bands from Top to Ten with stuff like JA2EMU TR8BA on 80m, KL7IRT and ZL4AV on 40m, and FP8, FR7, PZ5, TU2, ZD8 and 7P8 on the 15m band, again all s.s.b.

A letter from **Bernie Crockford** ZS1BW mentions a contest for amateurs and SWLs to celebrate the 150th anniversary of the University of Cape Town, from Saturday Feb 17 to Sunday March 4, 1979, with operation from 0600 to 2000 on Sundays, and 0700 to 1000 and 1500 to 2000 weekdays. Likely frequencies are 7050, 14210, 21200, 28580, subject to QRM. Contact/log ZS1UCT plus two other ZS1s for award. Details from: Awards Manager ZS1MO, PO Box 5100, Cape Town 8000, Rep. of South Africa.

Club Activity

Much more from the clubs this month than individual readers! **RAIBC** secretary, Frances Wooley G3LWY, reports that the Strumech tower raffled at the Leicester Show was won by disabled member Shirley Hesketh G4HES! Shirley has already coached to success several blind girls at Chorleywood College, for the RAE. Club net 3750kHz s.s.b. 1000 Tuesdays and 1400 Wednesdays. The Cheshire Homes net is on 3650 to 3700kHz on Thursdays 1330.

The **Silverthorn RC** meets Fridays 1930 at Friday Hill House, Simmons Lane, Chingford, London E4, with details from: C. J. Hoare, at that address. Newsletter "Spurious" would be considerably improved with less large cartoons and virtually useless photographs, and more details of events to come! Surely you must have a winter programme chaps? Tars Talk, journal of the **Torbay ARS** reveals that G3LHJ will give a slide show on Feb 24, with the society's annual dinner being held on March 10. Meetings are held at Bath Lane, Torquay (rear of 94 Belgrave Road). Details from: F. Bolton G3VTQ, 2 Lower Coombe Road, Blindwell Park, Kingsteignton, Newton Abbot, Devon.

Stevenage and District ARS continues to meet at British Aerospace, Gunnels Wood Road, Stevenage, Herts on first and third Thursdays at 2015. March 1 sees talk by GB3HR repeater group on proposed 23cm and 10GHz beacon, while the 10th sees a visit to the VHF Convention at the Winning Post, Twickenham, Middx. AGM is on the 15th. The **West Kent ARS** report in for the first time with details of meetings held at 2000 at the Adult Education Centre, Tunbridge Wells, on Fridays, with a junk sale on March 2. On March 30 an interesting talk by Tony Tory on microprocessors in amateur radio. Informal chit-chat and code practice on Tuesdays at the Drill Hall, Victoria Road, says Brian Castle G4DYF, 6 Pinewood Avenue, Sevenoaks, Kent, who will supply details. A new QTH for the **Swansea ARS** at the Sketty Park Sports and Social Club, Aneurin Way, Sketty Park, Swansea, on alternate Tuesdays, Feb 6 and 20 et seq. New members welcomed with open arms and a pint! Ring Peter Jones GW4GRI on Swansea 873986 for info or write to: 27 Gorwydd Road, Gowerton.

J. Bazley G3HCP, President of the RSGB, will be guest of honour at the 31st annual dinner of the **Sutton & Cheam RS**, taking place at the Woodstock Hotel on Saturday March 24. Details and tickets from: G. W. Brind G4CMU, 26 Grange Meadow, Banstead, Surrey.

Events to be held by the **Cheltenham AR Assoc** include: a constructors contest on Friday Feb 16 and a talk on i.c.s by Eric Hibbett G8LAY on Thursday March 1, with G4BSO and G3SSO discoursing on aerial planning permission on the 16th. Meetings at the Old Bakery, Chester Walk, Cheltenham at 2000. Info from: G. Martin G3IER, 88 Tennyson Road, Cheltenham, Glos.

Log Extracts (All s.s.b.)

B. Hughes:—10m HS1ABD KZ0DX P²8NKV S79MC STORK VS6FI VR3AH XT2AT 15m D68AD KJ6BZ TJ2AP VP2DAY 40m HR1HMV HK5BCI YS1RRD ZP5LX 4W1BC

I. Marquis:—10m HP1PJ VP2MBD ZL3AAX 15m CT2BB FP8DX FR7ZN PZ5AA TU2FH ZD8RG ZF2AG 7P8BC 20m VP2DAO ZE3JO 40m HK5DUS KL7IRT ZL4AV 80m FP8DX JA2EMU LX1ST TR8BA 9H1EU

Remember, all logs and letters by 15th of the month.





MEDIUM WAVE DX

by Charles Molloy G8BUS

It may be arguable, whether humility is the greatest virtue that a m.w. DXer should possess, but there is no doubt at all that selectivity is the attribute to look for in a receiver to be used for serious medium wave DXing. It is not that other factors such as sensitivity, stability, scale accuracy, freedom from overloading and cross-modulation, are unimportant. The present overcrowded state of the band, means that you should have the facility to winkle out DX that is close to a strong local station, and this means using a receiver that has good selectivity.

What is selectivity? It is the ability a receiver has to separate stations that are close to one another in the band. You can regard a receiver as a window into the frequency spectrum, the width of this window depending on the degree of selectivity. A narrow window means narrow (good) selectivity. It would be an easy matter, if all you had to do was to separate adjacent carriers, since a receiver with high selectivity would do the trick. Unfortunately the programme (modulation) requires space, the amount of space depending on the highest audio frequency. If it is 3kHz then a 6kHz bandwidth is required, i.e., 3kHz above and 3kHz below the carrier with the double sideband system currently in use. If bandwidth is reduced in order to reduce QRM then the audio range is reduced too, and in an extreme case speech becomes unintelligible.

How Selectivity is Measured

The handbook of a well-known receiver quotes the selectivity as 4kHz at -6dB and 18kHz at -40dB. What do these parameters mean? The first is the important one. 6dB is shorthand for 6 decibels and -6dB means "one half of the original value", so the signal at the sides of the 4kHz window is half that at the centre. Selectivity is invariably measured at the 6dB points. The statement 18kHz at -40dB is less important. It means simply that at 9kHz on either side of the carrier the signal will be 40dB down which is 1/100th of its original value. Ideally the signal would be zero at the sides of the window, but in practice it falls off gradually.

What to Look For

If you want to judge a receiver's ability to separate stations then look for the bandwidth at the 6dB points. It will be found under "Selectivity" in the handbook or specification. Sometimes it isn't mentioned at all so one can draw the appropriate conclusion. Occasionally on imported receivers the bandwidth is given as a plus or minus figure such as \pm 2kHz. Multiply by 2, as the real bandwidth is 4kHz! I recently studied a one-page "spec" for a receiver currently available and after some searching found Selectivity near the end. It was 4kHz at the 6dB points which is not very good for DXing. A few years ago I tried an experiment with CJON (now CJYQ) on 930kHz using a bandwidth of 2.4kHz. It could be heard easily between the Europeans on 926 and 935. When the bandwidth was increased to 5kHz, CJON just disappeared, as the stations on either side spread out to meet each other.

My BRT400 communications receiver has a sixposition selectivity switch which gives bandwidths of 0.5, 1.0, 2.0, 5.5, 9.0 and 13.0kHz at the 6dB points. Normally I use 2kHz when tuning around and this is increased to 5.5kHz or even 9kHz, QRM permitting, if I want to listen to the programme. A receiver with fixed selectivity must compromise between the needs of the DXer and the listener, and I would question whether such a receiver should be entitled to use the term "communications". A personal view that many will disagree with. It is a pity that the Q Multiplier, referred to recently by Eric Dowdeswell, has gone out of fashion as this simple device provides an easy means of obtaining variable selectivity, and it was incorporated in a number of moderately-priced receivers a few years ago.

If you do have a receiver with narrow selectivity then it is possible to use this facility and still hear the modulation. If the programme quality deteriorates as selectivity is increased, then detune slightly, away from the offending QRM. Speech quality will immediately improve, as your window will now look out on only one of the two sidebands. It may come as a surprise to some DXers to find out that the programme is actually carried twice on an a.m. double-sideband system. There is a sideband complete with programme on each side of the carrier and it is only necessary to tune to one of these to extract the modulation. You could double the number of channels on the medium waves by suppressing one sideband but there are problems in doing this.

The New Band Plan

I remained up late on the night of the big change over last November, just in case I might be missing something, and was well rewarded with the feeling that I was listening to history being made. After hearing the short announcement on 200kHz, at midnight, I tuned round the medium waves to be greeted by tuning notes on nearly every channel. The change at the h.f. end of the band was striking. Where the German power-house on 1602 had been, only a few minutes earlier, was now the third international common frequency with nothing to be heard but a burble. That evening I hunted around for any DX between channels, but all I could find was an unidentified Arab on 1570. It looks as if Asiatic DX has disappeared, in the evenings at any rate.

Readers' Letters

Steve Whitt (Cambridge) is interested in QSLs and he reports that he has had 100% returns from the 24 US stations he has reported to so far, but from further south, Radio Margarita on 1020 and Radio Coro on 1210 have not replied. This is a common experience as Latin Americans are notoriously difficult to QSL. One approach is to write a personalised letter to the station giving details of one's self, and perhaps a photo of the shack, or the locality. WEVD 1330 broadcasts in Hebrew, and Steve wonders how he can compile a report, as the programme material is meaningless to him. Make a tape of the DX and if you cannot get it translated then send the tape to the station. Many DXers send tapes instead of reports though it is rather an expensive way of doing it.

Two other questions come from Steve: 1. What is the station on 593 (pre-Geneva) underneath the West German that relays the BBC World Service? It is probably cross-modulation which can occur in the ionosphere as well as inside the receiver. If you listen on the open carrier of any

high power European, just before sign-off, you may hear this effect. There are so many megawatts floating around the ionosphere these days, that all sorts of weird effects can be observed. 2. What is the easiest English speaking DX from the Caribbean? None is easy. Try ZDK Antigue on 1100 or Radio Paradise, St Kitts on 1265, but you will have to stay up late to hear them.

DX Heard

John Faulkner (Mansfield) reports hearing 19 Canadian and 13 US stations with his Trio 9R59D receiver and 40 inch PW loop. Toronto was logged on three points on the dial with CBL on 740, CJBC 860 and CFRB on 1010. Others include WHAM Rochester, NY on 1180 and Fort Wayne, Indiana on 1190kHz.

An HMV domestic receiver and loop are in use in Birmingham by John Dennis Court and his log contains two not-so-often heard stations in Newfoundland; VOWR on 800 at 0030 and VOAR on 1230 at 0200. There are still a few outlets in Newfoundland that use the old prefix "V" in the call sign, VOCM on 590 being the one that is usually heard. Noel Cosgrave (Dublin) has a Mullard MAS1659 receiver and 36 inch loop, which brought him Radio Belgrano, Buenos Aires on 950kHz at 0255, Radio Tupi in Rio de Janeiro on 1280 at 0315 and CB57 Santiago, Chile on 570kHz.



SHORT-WAVE BROADCASTS

by Charles Molloy G8BUS

Skip is a term frequently used by DXers and it is one that may confuse the newcomer to the hobby. Very briefly it refers to a zone around a transmitter where little or no signal can be heard. The ground wave travels only a few miles from a short-wave station before petering out, while the nearest place where the sky wave returns to earth may be hundreds of miles away. The distance between the two is the skip distance. Anyone situated nearer to the transmitter than the point where the sky wave returns, will be in the skip zone and consequently will not hear the station at all, unless of course he is within range of the ground wave. This is the reason why BBC short-wave transmitters in this country are not heard within the UK. The sites of these transmitters incidentally are at Daventry in Northamptonshire, Skelton in Cumberland, Wooferton near Ludlow and Rampisham in Dorset.

The Sky Wave

Great stuff, you may say, but why is it that the sky wave fails to come back to the earth inside the skip zone. To understand this you have to look at what happens to a wave sent up vertically. If the frequency is low enough then the signal will travel some distance into the ionosphere before being returned to the transmitter. A higher frequency will go up a bit further before coming back, and if frequency is increased gradually, a time will come when the wave will travel right through the ionosphere and off into space. The highest frequency to be returned is called the critical frequency and is usually designated f_{e^*}

Vertical radiation is not much use for broadcasting so in practice a lower angle is employed. A wave travelling at an oblique angle will have to travel a greater distance through the ionosphere than a vertical one, before coming out at the top, so frequencies higher than the critical frequency can be used at low angles. The highest frequency that will be returned will be when the wave is at a low angle, just above the horizon, which is the case for long distance transmission and this frequency may be three or four times f.. The Maximum Usable Frequency (m.u.f.) for any particular angle can be calculated from f which in turn is fairly easy to measure. One final point. To get the maximum signal into the target area it is desirable to transmit as near as possible to the m.u.f., and any radiation at a higher angle than required for this distance will therefore penetrate the ionosphere and be lost. This is what happens to the signal that is missing in the skip zone.

Propagation and DXing

About 4000km is the maximum distance that can be covered by a single hop from low angle radiation and in this case the skip will be great. If the target is nearer than 4000km then a higher radiation angle will have to be used at the transmitter and a lower frequency will be needed as well. Obviously the skip will also be lower. Reception areas further away than 4000km will be reached by multiple hops, the wavefront being reflected from the earth's surface back into the ionosphere after the first hop. So, at any particular time of day, season of the year or period of the sunspot cycle, all of which affect the critical frequency, the highest frequencies available will be used for long distance working and lower ones for short distances. There is no use looking for your favourite European local on 13 metres, for even if the band is open the skip will be too great.

19 Metre Band (15MHz)

Following on from last month we will now have a look at 19 metres whose limits are 15100 to 15450kHz, though there is some spread on either side. 19 metres is mainly a daytime band with world-wide reception being possible. Look for Vietnam on 15012kHz, Teheran 15084, Japan 15105, RSA 15220, New Zealand 15130, Tanzania 15435. Some medium range DX can be heard during the day, such as Norway on 15175, Morocco 15195, Sweden 15240 and Finland on 15265, but these disappear as dark approaches and the m.u.f. falls.

During the evening Latin American DX can be heard and although at first sight this may appear surprising, it should be noted that the signal path is from the south-west from the southern hemisphere, where the greater part of the route will still be in daylight. DX heard regularly in the UK includes: Chile on 15 115, Brazil on 15 145, Chile on 15 150, Radio el Mundo Argentina on 15 290, Radio Nacional Colombia on 15 335, R. Mexico on 15 385, Venezuela on 15 400, and from the same area, Radio Grenada on 15 105kHz.

Readers' Letters

The MCR1 wartime receiver is mentioned again by **Trevor Goodenough**, who has two of them which he uses regularly with a 300ft long wire and an a.t.u. DX heard with this set-up includes Radio Japan and Radio

Reports on the various bands are welcome and should be sent direct, by the 15th of the month, to:

AMATEUR BANDS Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey KT21 2TW. Logs by bands, each in alphabetical order.

MEDIUM and SW BANDS Charles Molloy G8BUS, 132 Segars Lane, Southport PR8 3JG. Reports for both bands **must** be kept separate.

VHF BANDS Ron Ham BRS15744, Faraday, Greyfriars, Storrington, Sussex RH20 4HE.

Australia. Trevor is anxious to get hold of a copy of the circuit diagram or the manual for the MCR1 and if anyone can help, would they please contact Trevor direct at 8 Glencraig Terrace, Fenwick, Ayrshire. Postage will be refunded.

The logging of Radio New Zealand on 15 130kHz at 0630 by S.I. Fass in the December issue, is referred to by **T. W. G. Elsenham,** who points out that the RNZ schedule says that this channel closes down at 0450. Schedules are always changing and cannot be relied on. Mr Elsenham goes on to refer to the pre-war BBC weekly *World Radio.* This had a feature covering stations heard by readers which included details of the programmes, and he wonders if we could do the same here. Sounds a very good idea. There are some interesting programmes to be heard on the short waves, usually from the less conspicuous broadcasts. Details of programmes from DX stations should be brief and give the date, time, and frequency (if known).

DX Reported

An unusual log of 60 metre DX, all of Venezuelan stations, has been received from Leon R. Sin Sun of Aberdeen, who used a Sony CF9JOS between 2300 and 0400 to pull in Radio Lara on 4800kHz, Bolivar on 4770, Universo 4870, Juventud 4900, Yaracuy 4940, Rumbos 4970, Ecos del Torbes 4980, Barquisimeto 4990, Continente 5030. Programmes heard included local popular music, folklore and interviews which would be of interest to DXers who understand Spanish. Leon does not use any special aerial as he is rather short of space. Try the outside TV aerial if you have one or alternatively a Joystick plus a.t.u. might be useful, but consult the manufacturer, Partridge Electronics, who advertise in PW, to make sure that a Joystick can be used with your Sony receiver.

An FRG-7 plus 60ft loft aerial and a.t.u. are in use in Wigan by **Jim Edwards**, who reports hearing KGEI, La voz de Amistad, on 9575 at 1030, Baghdad on 9745 at 2130, Lagos Nigeria on 11 770 at 0630, WINB Red Lion on 15 185 at 2000. Jim reports that he is now a monitor for Family Radio WYFR, which he finds an interesting diversion when he is not DXing.

Australian Domestic Short-wave Stations

Lex Arnold, Hemel Hempstead (R107 ex-WD communications receiver and 9ft indoor aerial), sends some notes about the domestic service of the Australian Broadcasting Commission. He reports that ABC Melbourne has been heard regularly between 1000 and 1300 on 9680kHz. From past experience Lex suggests that DXers should try for stations in this network between late November and mid-April, especially during February and March. Listen for VLH9 on 9690 between 0730 and 1300, VLH11 on 11 880 from 1900 to 2115, VLH15 on 15 230 from 2230 to 0715, all from Melbourne; VLM4 on 4920 and VLQ 9660 between 2200 and 1400, which are in Brisbane; VLW6 on 6140 and VLW9 on 9610 from 2200 to 0100 and 1000 to 1600, both in Perth. Readers who think that Radio Australia is not really DX should try for some of these domestic stations. There is also Port Moresby in New Guinea with P2T9 on 4890 between 0715 and 1400 and again from 2000 to 2200 plus P2T9 on 9520 between 2215 and 0700.





First, our sincere congratulations to our reader **Robin Bellerby** G3ZYE, Hove, Sussex, on his election to the council of the Radio Society of Great Britain. Robin, a member of the Brighton and District Radio Society, the Mid-Sussex Amateur Radio Society and the Sussex Repeater Group, is actively interested in all bands from 160m to 70cm. News of his election victory came just after the Brighton and District RS had made Robin the first recipient of their trophy, the Bill Pitfield Memorial Award, given for meritorious service to amateur radio by a club member. Like all RSGB council members, Robin has some hard work ahead, what with society affairs, conferences, conventions and WARC 79, the outcome of which may well affect all of us. Congratulations also to **D**. **J. Stewart**, ex-G8MZP, on passing his Morse test and now sporting the call sign G4HSY.

Solar Activity

Nigel Fisher, Goff Gill, Robin Knight, Peter Mynheer and Chris Peeder have recently completed a 60MHz radio telescope for the radio section of the South-East Essex Astronomical Society, and were delighted when their new instrument, built with the r.f. and i.f. sections of an old TV receiver, recorded solar activity on December 10, 11 and 12, as did the 136MHz telescope of Cmdr Henry Hatfield, Sevenoaks, and the 146MHz receiver which I use. Henry recorded a slight increase in solar radio noise during the afternoon of November 24, and again on the 25th and, using his spectrohelioscope, he observed a large prominence on the north-west limb of the sun which lived for about 24 hours and he saw another on the east limb on the 26th.

It must have been this solar activity which caused the blackout on 80m, reported by **Alan Baker** G4GNX, Newhaven, at midday on the 25th, the aurora during the same afternoon and the ionospheric disturbance reported by the BBC World Service at 0215 on the 26th. At 1015 on the 25th, I did not hear any International Beacon Project signals on 10m and Alan said that on 80m, even the strong shipping stations which share the band were weak and, from Applecross, Western Australia, Anthony Mann writes: "November 26, absolutely dead, nil above 28MHz all day".

Aurora

The land lines were soon buzzing with an alert once **Roy Bannister** G4GPX, Lancing, John Cooper G8NGO, Cowfold, and **Dermot Cronin** G4GRO, Royal Sovereign, heard auroral signals on 2m. Dermot worked GI and GM and G4GNX, who was soon in on the action, heard PA3. **Dave Cox** G8OPR, Andover, Hants, worked stations in 5 different QRA locator squares during the event which included 3 GMs and a GI. For John Branegan GM8OXQ, Saline, Fife, this aurora was a novel event because, at 1135, he had a QSO with SM4IVE and soon after with G8OGD via OSCAR 8J. The SM thought an aurora was starting over central Sweden and the G8 told John that his signal had an auroral tone.

At 1320, John again contacted SM4IVE via OSCAR 8J, who reported that the auroral activity in Sweden and Finland was fading and thought to be heading towards Scotland, Between 1340 and 1400 John was receiving auroral signals from the Russian satellite, RS-1, and when he left the satellite he found a full scale aurora affecting the 2m band, which he monitored until 1840. During the period he heard tone "A" signals from the 2m beacons in Cornwall GB3CTC, France FX0THF, Germany DL0PR, Lerwick GB3LER, Northern Ireland GB3GI, and Wrotham GB3VHF, in addition to amateur stations from EI, G, GD, GI, GM and SM. GM8OXQ said there was another, but much weaker, aurora on the 26th but only the northern GMs beyond Aberdeen could use it. Anthony Mann says: "It would be lovely to witness an aurora here", and thinks that the only two visible there were during the 1946/7 sunspot maxima. It really is a wonderful sight Anthony, I will never forget the beauty of the one I saw from southern England, following a big solar event in August 1972.

The 10 Metre Band

M. Mrzyglod is delighted with the performance of his new FRG-7 and with his 10m ground-plane aerial, he has been logging the IBP signals from Bermuda VP9BA, Canada VE3TEN, Germany DL0IGI and s.s.b. signals from amateurs on both sides of the Atlantic. M. Mrzyglod is looking for a circuit which he thinks was published in *Practical Wireless* some years ago for a Band I converter. If anyone can help, please let me know. Neil Clarke BRS 34306, Nottingley, York has been enjoying the DX on 10m and logging A4XFA, OX3CO, and VS6FI.

Like myself, John Branegan has found that the signal from the Bahrain beacon A9XC, was very consistent between November 19 and December 15, in fact, I heard it every day except during the blackout on November 25. John, Alan Baker and I, also heard signals from 5B4CY, DL0IGI, N4RD, VP9BA and 3B8MS. On his newly acquired Eddystone 770R communications receiver, John can tune through Bands I and II, and has already observed the effect of F2 conditions on the American and European stations which operate between 30 and 40MHz. He is looking forward to using it during the coming sporadic-E season.

From Down Under

Anthony Mann reports strong sporadic-E disturbances on November 16 and 28, producing Band I colour on an indoor aerial and m.u.f. lapping Band II. On November 18: "The best F2 opening so far", says Anthony who received strong signals from the BBC, Channel B1 sound, 41.5MHz, French Channel F2 sound, 41.25MHz, a watchable picture from a South Korean station, strong Chinese video, 57.770MHz, strong Russian TV sound,

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east and west Malaysian video, and strong signals from Japanese amateurs in the 6m band. We are all very envious Anthony.

Tropospheric

During the morning of November 22, Alan Baker worked F1EZP and Mick Senior G4EFO, Horsham worked F6DOP, both through the Brighton repeater GB3SR, R3, and for most of the 22nd and 23rd I heard GW stations working through GB3BC, R6. At 2201 on the 19th, Alan had an unusual contact which lasted 28 minutes, through GB3SR; he worked N2AFE/MM, Ralph, from 5th Avenue, Brooklyn, New York, who was in the engine room of the *Export Patriot* in the English Channel, and as soon as other people realised what was on, suddenly Ralph had 19 stations after him.

A more extensive v.h.f. opening took place on December 5, 6 and 7, during which time I received strong signals from both the Bristol Channel and Dover 2m repeaters. Throughout the 6th I received good pictures from the IBA transmitter at Lichfield, Channel 8, 189MHz, and signals from several continental broadcast stations in Band II. At 0030 on the 6th. John Cooper phoned G4GNX to say that he had worked a station through a Belgian repeater, and G4GNX reported hearing ON0OV on R4. During the morning, GU2FZC, St Peter Port and F1EBE, Rouen, worked through the Brighton repeater and in the afternoon GB3BC was putting a strong signal into east Sussex. Periodically, throughout the day, Band V TV suffered from co-channel interference and in the evening, G6GL, Radlett, Hertfordshire, had an effortless QSO with G3TIR in Crawley, Sussex, via the Hampshire repeater, GB3SN, R5. Signals from the French repeater on R9 were heard on December 8 by G4GNX and G4GPX, and G3TRY, High Wycombe worked a station in Yorkshire through extreme QSB.

Satellites

On December 1, our satellite expert, John Branegan, wrote: "Being a thorough optimist I still go on OSCAR 8A despite the high m.u.f., and though the ionosphere must be as thick as treacle, I have twice got across to W2BXA in New Jersey". The W2 told John that he was the only European he had heard for some time. "On Mode J at weekends it is a very different story", says John, who in 10 QSOs in November, worked 4 stations, W9KDR Conn, WB2OXJ New Jersey, WA3ZHW Penn, and VE2LI Montreal. By the end of November, John, by working his first east-German, DM2DIN, s.s.b., made his score 27 countries on 8J and pushed his total up to 35 countries via satellite.

Contests

As usual the RSGB have catered for the v.h.f./u.h.f. enthusiasts in their contests calendar for 1979, and reports will be welcome from any of our readers who take part in either the transmitting or receiving sections of any of the following events:

	. 144/432MHz and SWL
•	. 432MHz open and SWL
	. 70MHz Open
	. 144MHz Portable
	. VHF NFD
•	. 70MHz Open
•	. 144MHz Open and SWL
	. 70MHz Fixed
	. 144MHz Fixed
	• • • •



by RON HAM

"THE TWO ERNS"



One of the fascinations for me when writing my column is to realise that my readers are using wavelengths from 10 metres through to 3 centimetres, which in frequency terms, is 30 million to 10 thousand-million hertz. It is men like Ern Downer G8GKV of Worthing (right), and Ern Hoare, G8BDJ of nearby Southwick (left), known locally as "The Two Erns", who are pioneering the microwave end of the radio frequency spectrum.

Ern Downer's schoolboy interest in amateur radio was fostered by a neighbour, the late G2XO of London. During the early 1930s Ern built himself a number of short-wave receivers, from designs published in the amateur radio press. His introduction to v.h.f. came in the second world war with the Royal Tank Regiment. After the war he spent some years in East Africa where he used "point-to-point" links, and on his return to the UK, a colleague, G3YHM, invited him to join the Worthing Amateur Radio Club. In May 1972 he took the RAE, and by August, his call, G8GKV, was heard on the air. In 1927, the 10-year-old Ern Hoare began building

In 1927, the 10-year-old Ern Hoare began building crystal sets, progressing to valve sets, and later to a 30-line scanning disc television receiver, through which he did get a picture of sorts. Ern was a radar operator with the Royal Artillery during the second world war, using both v.h.f. and microwave systems. Later, he volunteered for the 6th Airborne Division, where he used the famous 38 sets for 3in mortar fire control. He took the RAE in 1962 and was first licensed as G3RZD/T, constructing his own gear for amateur TV. When his call was changed to G8BDJ he used home-brew gear on 2m, 70cm and 23cm.

Due to common interests, both G8GKV and G8BDJ began to inhabit the local hill tops, operating portable stations on 2m. When 'BDJ began building equipment for 10GHz, 'GKV was hooked, since when "The Two Erns" have never looked back. Now, with some very impressive home-brew portable stations, they each hold many well deserved awards, and apart from entering the RSGB's 10GHz contests, they spend a great deal of time trying out "difficult" paths of microwavelength communications.

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7401	14p	74100	130p	74259	250p	74LS195	140p	74C161	155P	AY1-1313	968p	*MC3340	120p	AD149	70P	BFY56 33p	TIP2955	780	*2N4036	65p	*0A81	15p
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160V:0.039,0.15,0.2211p;0.33,0.4719p;0.68,1-0.22p;1.529p;2.232p;4.736p, DUBLIER: 1000V: 0.01,0.015 20p; 0.022 22p; 0.047 26p; 0.138p; 0.138p; 1.0 175p, POLYESTER RADIAL LEAD (Valuasin //i. 250V: 0.01,0.016,0.22,0.0275p;0.033,0.047;0.068,0.17p;0.1511p; 0.02,0.3313p;0.4715p;0.6818p;1.024p;1.527p;2.231p 1000pF/350V 8p ELECTON THE CARACTORS	AF116* 30 AF117* 30 AF118* 55 AF121* 48 AF124* 55 AF125* 35 AF126 50 AF127* 35	BC4461* 36 BC477* 25 BC547 12 BC548 12 BC549C 13 BC558 16 BC558 16	BFR40 25 BFR40 25 BFR41 28 BFR79 28 BFR80 28 BFR80 28 BFR81 28 BFR98 105 BFX29* 28 BFX29* 24	0C77* 66 0C79* 76 0C81D* 28 0C82D* 40 0C83* 48 0C84* 44 0C122* 48	ZTX500 15 ZTX501 15 ZTX502 19 ZTX503 15 ZTX504 25 ZTX5504 25 ZTX5504 25 ZTX550 25 402501 85	2N3615* 135 2N3663* 26 2N3702 11 2N3703 11 2N3704 11 2N3705 11 2N3705 11 2N3705 11
ELEV. 10 440 it 127 637 2500 it 5:048 6330 it 6:43 16 22, 2.5, 3, 4.7, 6, 8, 8, 10 15, 22, 847, 32, 551 176, 83, 100, 276, 550 v; 10, 79, 550, 100, 220 259, 470 506, 1000 62, 400 v; 22, 33 it 7 16, 100, 116, 220 6300 520, 4700 6469, 350 v; 10, 33 70, 500 276, 1000 4690; 250 v; 10, 22, 47 69; 500, 100, 160 86; 220, 250, 136, 1470, 640 259; 1000 276, 1500 309; 2200 419; 3300 559; 4700 549; 160 v; 10, 40, 47, 66 79; 100, 120 259; 200 330 149, 470 159; 1000, 1500 259; 2470 549; 100 v; 10, 66 79; 1000 149, 740 700; 15, 000 4590; 250 v; 4700 459; 200 349; 100, 200 + 2000 959.	AF139* 35 AF178* 70 AF180* 70 AF239* 42 AF212 126 ASY26* 40 ASY27* 45 ASY50* 95 ASY76* 95	BCY30* 57 BCY30* 75 BCY39* 80 BCY40* 78 BCY40* 78 BCY42 48 BCY42 48 BCY58 22 BCY59* 22	BFX85* 28 BFX86* 28 BFX86* 28 BFX88* 28 BFX88* 28 BFY18* 50 BFY50* 20 BFY51* 20 BFY52* 20	OC139• 110 OC140• 110 OC141• 110 OC170• 40 OC171• 40 OC200• 48 OC201• 76 OC202 95	40251* 97 40311* 50 40313* 125 40315* 55 40316* 85 40317* 52 40319* 71 40320* 56	2N3708 11 2N3709 11 2N3710 16 2N3711 12 2N3771 12 2N3772* 195 2N3773* 288 2N3819 22 2N3820 45
TANTALUM BEAD CAPACITORS POTENTIOMETERS (AB or EGEN) 35V: 0-1µF, 0-22, 0-33, 0-47, 0-68, 1-0, 2-2µF, 3-3, 4-7, 6-8, 25V: 1-5, 10, 20V: 1-5µF 13p each. 10V: 22µF, 33, 6V: 22µF, 47, 68, 3V: 100µF 16p each. 10V: 100µF POTENTIOMETERS (AB or EGEN) Carbon Track, 0-25W Log & 0-5W Linear Values. 5000, 11k 8, 2K (LIN ONLY) Single 27p 35p, 16V:47, 100µF 40p. 27p 5K0-2MQ aingle gang D/P switch 60p 27p 5K0-2MQ aingle gang D/P switch 60p 27p 6K0-2MQ aingle gang D/P switch 60p	BC107* 9 BC107B* 10 BC108* 9 BC108B* 10 BC108B* 10 BC109B* 12 BC109B* 12 BC109C* 12	BCY70* 15 BCY71* 20 BCY72* 20 BCY78* 25 BCZ10 145 BCZ11 145 BD112 95 BD115* 65	BFY53* 28 BFY55* 45 BFY64* 40 BFY71* 20 BLYB3* 649 BSX20* 18 8SX20* 18 8SZ26* 75 8SX29* 45	OC203* 85 OC204* 85 SJE5039* 95 TIP29 43 TIP29A 44 TIP29B 56 TIP29C 60 TIP30 47	40323* 60 40324* 85 40326* 52 40327* 62 40347* 80 40348* 105 40360* 43 40361* 45	2N3823* 65 2N3824* 70 2N3866* 90 2N3903 20 2N3904 18 2N3905 18 2N3905 17 2N4037* 52
TUDAL FILM CAP 2015 Output Stips 100V:001.0052.001.0050.0050 005.0050 005.0050 005.0050 0.1pf.002.003.004.005.0056 pf 7p 0.25W log and linear values 60mm track 0.1pf.002.9p. 50V:0.47 pf 12p CERAMIC CAPACITORS 50V 7p 10K0-500K0.004 gang 80p Fange:0.50F to 10nF 3p 15nF, 22nF.33nF, 47nF 4p 100nF 6p 0.1W500-22 MMIN.Vert. & Horiz. 8p	BC113 20 BC144 20 BC115 20 BC116 19 BC117 20 BC118 20 BC118 20 BC119* 28 BC134 20	BD123* 98 BD124* 115 BD131* 45 BD132* 45 BD132* 43 BD135* 38 BD136* 36 BD136* 36	BSX78* 55 BSY95A 18 BU105* 140 BU205 190 BU208 228 E421 96 MD8001 158 ME1120 25	TIP30B 64 TIP30C 65 TIP31* 40 TIP31A* 40 TIP31A* 40 TIP31C* 65 TIP32* 55 TIP32A* 55	40406* 65 40407* 52 40408* 75 40411* 295 40412* 65 40467* 95 40594* 90 40595* 98	2N4055* 17 2N40655* 17 2N4062 17 2N4064* 120 2N4069 45 2N4236 145 2N4286 20 2N4289 20
POLYSTYRENE CAPACITORS: 10pF to 1nF, 8p. 0-25W 100C-3 3 MQ Horiz. 0-25W 250Q-4-7MQ Vert. larger 10p 10p SILVER MICA (pF) 8.5	BC135 20 BC136 18 BC137 20 BC140* 35 BC142* 30 BC143* 30 BC147 8 BC1478 10	BD138* 50 BD139* 40 BD140* 36 BD142* 59 BD144* 198 BD145* 198 BD145* 198 BD181* 35 BD205* 110 BD222 60	ME4102 10 ME6002 10 MJ400* 90 MJ491* 160 MJ2955* 95 MJE340* 54 MJE370* 58 MJE371* 60	119328* 70 T1932C* 75 T1933* 80 T1933A* 80 T1933B* 100 T1933C* 110 T1934C* 85 T1934A* 85 T1934B* 110	40603* 65 40636* 125 40673* 68 2N697* 25 2N698* 44 2N699* 54 2N706A* 19 2N707* 39 2N707* 19	2N4427* 75 2N4859 65 2N4922* 55 2N5135 42 2N5136 42 2N5138 20 2N5172 25 2N5179* 60 2N5180* 60
Bit Days Computer Name Computer Name	BC148 8 BC148 10 BC149 8 BC149 10 BC153 27 BC155 10 BC156 11 BC159 11 BC169 11 BC167 10 BC168 12 BC168 13	BD376* 65 BD4374 45 BD517* 65 BD695A* 65 BD095A* 65 BDY17* 195 BDY56* 156 BDY56* 110 BDY61* 110 BDY61* 110 BDY61* 110 BF115* 34 BF115* 25 BF158 29 BF160 30 BF161 60	MJE221* 65 MJE221* 67 MJE295* 99 MJE3055* 70 MPF103 36 MPF103 36 MPF105 36 MPF105 50 MPF107 50 MPS3904 40 MPSA06 25 MPSA06 25 MPSA02 42	Tip34C* 110 Tip35A* 219 Tip35A* 225 Tip35B* 240 Tip35C* 270 Tip36* 280 Tip368* 280 Tip368* 280 Tip368* 280 Tip368* 280 Tip368* 285 Tip41A* 66 Tip41A* 68 Tip428* 82 Tip235* 65	2N914-32 2N916-327 2N918-40 2N920-51 2N1131-22 2N1131-22 2N1303-50 2N1304-50 2N1304-50 2N1305-32 2N1305-35 2N1307-50 2N1308-45 2N1670-150 2N1670+152 2N16711-215	2N5191*70 2N5305*40 2N5305*40 2N545732 2N545932 2N545932 2N548535 2N5642*750 2N577740 2N602740 2N602740 3N128112 3N140112 Matched Pair 10pextra
OUC 2355pr 275p 00-3825pr 430p TRIMMERS mini 100kHz 385p 100kHz 385p 2:5pf;3:10pf; 1040pf 22p 100kHz 385p 1080kHz 385p 5:25pf;6:5pf;85pf 88pf 30p 1MHz 323p 709C 8 pin 35 LM301A* 30 NE5 6:000 fb 45pf;3:10ef; 1MHz 325p 709C 8 pin 35 LM301A* 30 NE5 5:40pf; 10:80pf 22p 166MHz 385p 741*8 pin 25 M31B* 205 S3 25:200pf 33p 166MHz 335p 741*8 pin 36 M33B* 80 SN 100-500pf 45p 3-2768MHz 323p 753 8 pin* 150 LM34B* 95 SN 8Uzzers 6V or 12V 85p* 4-0324Mz 323p A13-10159 LM349* 325 SN 1µf, 4.7, 10, 22, 47, 100, 5-0MHz 355p AY-1-0212, 580 LM381NA 248 SN 355 SN 5,00mH 35p	370TBA 450 771 450 1136D 220 1236D 220 741 235 742 235 76003N 140 76013N 140 76013N 140 76013N 140 7603N 744 7603N 120 740 7603N 76033N 120 74 76033N 76033N 120 741 741 76013N 140 741 741 76033N 120 741 741 76115N 215 741 741	FL74* (TEX , 10 13 7475 11 13 7476 12 14 7480 13 14 7481 14 7481 15 18 7482 15 18 7484 16 38 7484 17 7486 17 7486 19 17 7486 19 17 7486 19 17 7486 10 16 74890 10 16 74890 11 7489 12 74892 13 7482 14 7482 14 7482 15 7482 16 7482 17 7482 18 7482 18 7482 18 7482 18 7484 18 7482 18 77486 18 77486 18 77486 18 77486 18 77486 19 77486 18 77486 18 77486 18 77486 18 77486 19 77486 10 77486	AS) 74153 64 74154 96 88 74155 53 36 74155 53 36 74155 53 36 74155 53 36 74155 53 36 74157 65 86 74163 92 95 74163 92 210 74163 92 210 74163 92 210 74164 105 105 75 74166 140 75 74166 140 75 74166 140 76 740 720	74265 63 40 74273 320 40 74278 440 40 74279 119 40 74284 373 40 74284 385 40 74283 125 40 74293 125 40 74293 125 40 74293 236 40 74293 236 40 74283 385 40 74284 385 40 74285 395 10	21 95 4072 22 85 4073 23 22 4075 24 66 4076 19 4077 26 180 4077 26 180 4078 29 99 4085 30 99 4085 30 58 4086 31 205 4089 32 100 4093 33 145 4094	21 4507 55 21 4508 299 23 4510 99 85 4511 150 40 4512 98 21 4513 206 20 4514 265 20 4514 265 20 4514 265 73 4517 382 73 4517 382 73 4519 55 190 4520 122
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20V-1.2A; 25V-1A; 30V-8A 30V-8A CA3028A* 80 MC3360P 120 TBA 3B0p (50p.p8p) CA3035 240 MC3401 52 TBA 100VA: 12V-4A 12V-4A; 15V-3A 15V-3A; CA3036 110 MK780 205 TBA 20V-2: 5A; 20V-2: 5A; 30V-1: 5A 30V-1: 5A; CA3036 100 MK780 205 TBA 40V-1: 25A; 40V-1: 25A; 30V-1: 5A 30V-1: 5A; CA3045 140 MK760040* 97 TDA (60p.p8p) CA3045 140 MK50398* 650 TDA CA3045* 71 MK50398* 650 TLO (N.B. P8 P charge to be added above our normal poteit charge. CA3075 75 MM57160 620 TLO DENCO COILS CA3081* 190 NE543X 210 NE543X 100 NE544	810s 95 74 8200 70 74 92002 260 744 9254 120 744 1022 575 744 20203 200 744 1022 575 744 61C* 76 744 64CP* 125 744 64CP* 125 744 64CP* 76 744 71CP* 76 744	30 74119 33 74119 38 33 74120 10 17 74121 11 74 74121 12 68 74123 13 115 74125 14 12 74126 15 94 74128 16 94 74132 17 82 74136 18 56 74136	149 74185 135 115 74186 275 25 74186 275 46 74190 95 48 74191 95 38 74192 98 38 74193 98 74 74193 98 73 74195 98 65 74196 93 56 74197 80	4000 15 40, 4001 15 40, 4002 16 40, 4006 93 40, 4007 18 40, 4008 82 40, 4008 82 40, 4010 38 40, 4011 18 40, 4011 18 40,	48 58 4411 1 49 48 4412 5 50 48 4415 5 51 72 4415 5 53 72 442 5 54 110 4433 5 57 2570 4440 1 59 480 44450 1	040 4543 155 1380 4549 375 795 4550 398 795 4556 72 545 4557 451 995 4558 105 825 4559 375 275 4560 210 295 4561 65
VALVE TYPE RDT2 92p CA3090AQ 338 NE595- NE556DB 20 TLO Ranges: 1-5 BLY, RFC 5chokes 91p CA3123 200 NE556DB 60 TLO Ranges: 1-5 BLY, RFC 5chokes 91p CA3130* 85 NE560* 325 TLO G. Wht. 86p RFC 7(19mH) 96p CA3130* 85 NE560* 325 TLO 1-5 Green 92p 17. 85p ICL7106* 752* NE5624* 425 UAA Tuning). TFT 18/1 6/ 16 or 4653 95p ICL7107* 975 NE564* 425 UAA Ranges: 1-5 BLY, MW/EFR 82p ICL8038CC* 340 NE566* 160 Z44 Rd.Wht. S2p MW/LW5FR 103p LD130* 452 NE567* 170 ZN4	1/2CP* 125 74 74CN* 199 74 81CP* 48 74 82CP* 96 74 84CP* 130 74 84CP* 130 74 174 90 74 125 4130 74	50 17 74142 51. 17 74143 53 17 74143 54 17 74144 50 17 74145 50 17 74145 50 28 74145 50 28 74148 73 32 74150 74 27 74151	209 74198 150 314 74199 150 314 74221 132 65 74246 204 175 74247 204 109 74248 240 99 74248 240 99 74249 204 64 74251 125	4013 42 40 4014 80 40 4015 82 40 4016 45 40 4017 82 40 4018 87 40 4019 48 40	30 115 4451 33 110 4452 56 58 4490F 57 380 4490V 58 22 4501 39 20 4502 70 32 4503 71 21 4506	295 4562 533 4566 155 695 4568 268 525 4569 280 19 4572 26 120 4580 594 695 4581 297 51 4585 101

WATFORD ELECTRONICS			OPTO ELECTRON LEDS+Clip TIL 209 Red 13	CS- 7 Segment Displays	SWITCHES. TOGGLE 2A 250V	NEW CONTRACTOR		VDU Chip	
Continued	from opposit	te side)	TIL211 Grn 18 TIL212 Yellow 22 0-2" Red 15	TIL307 675 TIL321 5"C.An 115	DPST 34 DPDT 38 4 pole on off 54		7	and MODU	LE
DIODES AA119 15 AA129 25 AA120 25 AA215 15 BA100 10 BY100 24 BY126 12 BY127 12 CR033 148* 0A3 75 0A47 12 0A79 12 0A81 15 0A85 12 0A90 6 0A91 6 0A920 9 0A200 9 0A202 8 Nay14 4	*BRIDGE RECTIFIERS (plastic case) 1A/50V 20 1A/20V 20 1A/20V 25 1A/400V 25 1A/400V 35 2A/100V 44 2A/400V 44 2A/400V 45 2A/100V 44 2A/400V 53 2A/600V 65 2A/100V 75 4A/400V 75 4A/400V 75 4A/400V 75 6A/200V 78 6A/400V 85 8V164 56 VM18 DIL 40 EXENSES B012 0/72 30V	SPEAKERS S0 3W 2':2' 65 2:5':3' 4002.5'' 58 55W 7''*4'' 190 8C33W 6''x4'' 6''x4'' 160 TRIACS* 3A/100V 8A/300V 8A/400V 516A/400V 16A/100V 95 16A/400V 95 16A/20V 95 16A/20V 95 16A/20V 93 940669 95 940609 95 95 96 972 <	0.2: Yellow, Grn, Amber 19 ORP61 M42 ORP12 63 2N5777 63 OPT0 15 OPT0 1	ILigg2-5. Cifn 113 DL704-3": Cith 99 DL707-3": Cith 99 MAN3640 165 XAN351-3" Green 180 Liquid Crystal Display 3½ OR 4 digit 915p TORS* 1 Plastic (TO220) case / -ve 0-5A: 5V. 6 2V. / 8:2, 12, 15V 76 18: L12, 15V 76 19: LM320-15ve 165 18 19: LM320-15ve 199 Variable Type 723 - 2 to +37V 45 e LM304H0te -400 2400 LM317K + 1: 2-37 100 LM317K + 1: 2-37 100 M117K + 1: 2-37 124 100 LM326N ± 152 240 LM326N ± 12V 240	SUB-MIN SUB-MIN TOGGLE SP changeover 59 SPSTon off 54 SPSTbiased 85 DPDT 6 tags 70 DPDT 6 tags 70 DPDT 6 tags 70 DPDT 6 tags 70 DPDT 8 and 115 SLIDE 250V 1A DPDT 14 1A DPDT 20 A DPDT 20 A DPDT 20 A DPDT 20 A DPDT 6 tags 85 DPDT 6 tag 85 DPDT	"PURBECK OSCILLOSCOPE" AUTHOR APPROVED PARTS including PCBs available.		VDU by using the r Thompson-CSFTX controller chip, SF. 16 line by 64 Charet text refreshment. Curs Management, Curs Management on sc Line erasing. Comp with any Computin system. SF.F 96364E AY-3-1015 AY-5-1013 UART SFC71301 ROM SFS80102 RAM 74LS163 UHF Modulator Compl. Module Data Booklet	to a ew /-CRT F96364, icters iursor or reen. atible g f11.75* f5.60* f2.05* f1.18* 250p* f136-50* f136-50* a30p
N916 5 N400/2* 5 N4003* 6	400mW 9p Rng: 3V3-33V 1-3W 17p	0-1 (cop 2 ¹ / ₂ ×3 ³ / ₄ " 46p	0-15 0-1 0 per clad) (plain) 39p 31p 2	15 ROCKER: (white) 1 over centre off Ap ROCKER: (illumin	5A 250V SP change- 30 hated, red) Chrome	Cathode Ray Tube 3BPI £7 (p&p insured 90p).	7.75*	ASCII Coded 56 Ke Keyboard	ey 0/75558-1
N4004/5* 6 N4006/7* 7 N4148 4 S44 20 3A/100V* 18 3A/400V* 20 3A/600V* 27 3A/1000V* 30 5A/600V 65	VARICAPS MVAM115120 BA102 25 BB104 40 BB1058 40 BB106 40 Noise Diode 25J Z5J 160	21×55 31×31, 55p 31×57 21×17, 169 31×17, 218 41×17, 218 41×17, 280 Pkt of 36 pins Spot face cutter Pin insertion tool VERO WIRING Spare Wire (Spoc	50p — 3 50p — 3 67p 50p 4 p 135p — 5 p 180p 141p 1 p — 1 p — 1 p — 1 p — 1 p — 1 1 2 5 2 2 2 2 2 2 2 2 2 2 2 2 2 2 2 2	Ip Beddi SA 220V SP RotARY: "Make-A. Wake-A. yourown muttiway S Stop Shafting Assem 20 Accommodates up to 30 Mains Switch DPST 5p Brop Shafting Assem 20 Accommodates up to 5p Brop Switch DPST 5p Brop Sylve Wake Way, 3p 5p Spacer and Screen P Pay Y, and Screen	52 Switch ' Make witch. Adjustable bly, 6 Wafers. 69 o fit 34 /afers. 1 pole/ /4 way. 4p/3 way, 5 blo Seen	Send S.A.E. for leaflet. 74Lso (TEXAS) 74LS01 33 74LS47 63 74LS01 33 74LS47 63 74LS01 31 74LS48 120 74LS02 14 74LS49 120 74LS02 14 74LS49 120 74LS02 14 74LS49 120 74LS02 14 74LS56 30 74LS03 12 74LS36 30 74LS03 27 74LS36 30 74LS03 20 74LS03 20 74LS04 20 74LS10 20 74LS	74LS173 74LS174 74LS175 74LS181 74LS183 74LS189 74LS190 74LS191 74LS192	105 74LS251 134 106 74LS253 142 110 74LS253 142 138 74LS257 110 398 74LS258 110 298 74LS258 160 74LS261 450 140 74LS266 52 140 74LS273 244 132 74LS273 25 550	4LS353 228 74LS365 65 74LS366 65 74LS367 65 74LS368 66 74LS373 180 74LS375 160 74LS377 212 74LS378 184
SCR'a* Thyristors 1A50V 38 1A100V 42	ALUM.BOXES with lid 3x2x1" 48	FERRIC CHLOI 11b bag Anhydro DALO ETCH RE Plus spare tin	RIDE* aus 65p + 30p p. & SISTPEN*	p. 2 to 4 way, 4 pole/2 to ROTARY: (Adjuste 1 pole/2 to 12 way, 2 2 to 4 way, 4 pole/2 to ROTARY: Mains 256	2p/2 to 6 way, 3 pole/ 3 way 41 0VAC, 4 Amp 45	74LS12 23 74LS75 48 /4LS14/170 74LS12 23 74LS76 40 74LS148 173 74LS13 38 74LS78 40 74LS151 96 74LS14 75 74LS83 115 74LS153 76 74LS15 30 74LS85 118 74LS153 76	74LS193 74LS194 74LS195 74LS196	130 74LS279 66 7 166 74LS280 250 1 136 74LS283 192 1 100 74LS290 128 7	74LS379 215 74LS384 86 74LS385 74LS386 86
A200V 47 A400V 52 A600V 70 A300V 35 A600V 70 A300V 35 A600V 70 A300V 43 A500V 58 A600V 58 A600V 59 12A300V 59 2A500V 92 2106D 38	$\begin{array}{c} 24\times51\\ 4\times42\\ 4\times24\times1\\ +& 68\\ 4\times24\times1\\ +& 64\\ 4\times51\times1\\ & 64\\ 4\times24\times2\\ & 84\\ 5\times4\times2\\ & 82\\ 5\times4\times2\\ & 88\\ 7\times5\times2\\ +& 14\\ 10\times7\times3\\ & 148\\ 10\times7\times3\\ & 148\\ 10\times4\\ 1\times3\\ 12\times5\times3\\ & 162\\ 12\times8\times3\\ & 250\\ \end{array}$	COPPER CLAD Fibre Single Glass sided Glass r35p 6"×12" 130p SOLDERCONP 100 pins 50p: DIL SOCKETS 8 pin 10p: 14 pin 20p; 20 pin 27p; 28 pin 42p; 40 pin 27p;	BOARDS* Double-SR sided 7.5" 90p 6 175p 7 1000 pins 40 •: Low Profile (TEX) 12p; 16 pin 13p; 18 12 22 pin 30p; 24 pin 31 n 55p; 64 pin 220p	PW PROJECTS BP General Coverage RE 24hrs. Digital Clock, Organ. General Purpt & Smoke Sensor A "PURBECK" Oscillosc Meter, "AVON" 2: Acoustic Delay Line. Send SAE plus 5p per	cciver, Chromachase, 'JUBILEE' Electronic ose SW Receiver, Gas Iarm, Metal Locator, ope, Audio Distortion n FM Transmitter, list.	74.513.00 74.580 116 74.515 56 74.520.20 74.580 38 74.515 56 74.520.20 74.581 38 74.515 56 74.522 74.531 38 74.551 56 74.522 74.531 38 74.551 56 74.527 74.531 38 74.551 56 74.527 74.531 38 74.515 96 74.527 74.531 38 74.551 98 74.527 74.530 27 74.551 98 74.532 74.510 56 74.512 98 74.533 74.510 56 74.516 16 74.533 74.510 57 74.516 16 74.533 74.511 50 74.516 16 74.537 39 74.511 50 74.516 16 74.533 39 74.511 50 74.516 156 74.537<	74LS197 74LS200 74LS200 74LS202 74LS221 74LS241 74LS241 74LS243 74LS243 74LS245 74LS245 74LS245 74LS245 74LS247	140 74L5293 128 ; 74L5295 185 ; ; ; 348 74L5295 185 ; ; 347 74L5298 681 ; ; 347 74L5298 681 ; ; 347 74L5294 681 ; ; 348 74L5300 175 ; ; 326 74L5302 175 ; ; ; 326 74L5302 175 ; ; ; ; 326 74L5302 175 ; <td>4L \$390 230 4L \$395 218 4L \$395 218 4L \$396 215 4L \$399 276 4L \$399 230 4L \$447 144 4L \$447 144 4L \$468 182 4L \$669 182 4L \$669 182 4L \$667 3 1050 L \$674 1450</td>	4L \$390 230 4L \$395 218 4L \$395 218 4L \$396 215 4L \$399 276 4L \$399 230 4L \$447 144 4L \$447 144 4L \$468 182 4L \$669 182 4L \$669 182 4L \$667 3 1050 L \$674 1450

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Published on approximately the 7th of each month by IPC Magazines Limited. Westover House. West Quay Road. POOLE. Dorset BH15 IJG. Printed in England by Chapel River Press. Andover. Hants. Sole Agents for Australia and New Zealand—Gordon and Gotch (Asia) Lid.; South Africa—Central News Agency Ltd. Subscriptions INLAND and OVERSEAS £10-60 payable to IPC Services. Oakfield House. Perrymount Road. Haywards Heath, Susses. PRACTICAL WIRELESS is sold subject to the following conditions. namely that it shall not, without the written consent of the Publishers first having been given be lent. resold. hired out or otherwise disposed of in a mutilated condition or in any unauthorised cover by way of Trade or affixed to or as part of any publication or advertising. literary or pictonal matter whatsoever. 79

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Presented free with Practical Wireless, March 1979

Finishing it off

Give your project a really professional appearance

A growing number of home constructors seem to be aiming to finish their projects in a more professional manner. There is a growing interest in using better looking and better finished cases and cabinets and the days of the old tobacco tin seem, fortunately, to have passed.

The choice of cases available to the home constructor has improved in recent months and several companies have made some of their professional cases available on the amateur market.

In most projects presented in *Practical Wireless* the decision as to the type and style of case to be used has already been taken. A lot of thought goes into the choice of case for each project, bearing in mind such parameters as price, the use to which the project will be put, availability and styling.

Simple Boxes

Some simple projects can be adequately housed in a simple plastic box if electrical screening is not required, or in the traditional diecast aluminium box if it is needed. Plastic boxes have been developed by several makers into quite sophisticated units offering many advantages for the amateur user. They are relatively inexpensive and can be easily drilled for controls using only simple hand tools, although some degree of care is needed if the smooth polished surfaces are not to be marked. A wide range of sizes and styles are available and several types have mounting facilities for printed circuit boards moulded into the box, simplifying construction.

The diecast box in its simplest form has been around for several decades andthere are probably more pieces of electronic equipment built into one of these than into any other type, with the possible exception of the infamous tobacco tin. It has several disadvantages for the amateur, being quite difficult to work on with simple hand tools and having a rather indifferent appearance in the 'as bought' state. Versions are available with a respectable gloss painted finish, but these are very expensive and unless the electrical screening properties of the diecast box are really needed then the plastic versions are better, and cheaper.

The plain simple aluminium box with removable lid offers a stark and cheap housing without any pretensions as to style or elegance.

Test Gear

When it comes to projects such as test equipment, domestic audio or amateur radio projects, some respectable form of housing becomes essential.

Test equipment should be built to be used, and must offer reliability and confidence and it cannot really give of its best housed in a piece of bent aluminium. The case design chosen must be mechanically robust, easily worked and allow the controls to be ergonomically positioned. If the instrument is intended to be used in the workshop then allowances should be made for maximum utilisation of bench or shelf space. Take a look at what the commercial instrument makers use and don't be ashamed to take the best ideas from all of them. Don't be tempted to use a case that is so small that the controls have to be cramped together making the instrument awkward to use. You should be proud of the finished instrument and not be ashamed to leave it permanently on the workshop bench where it will get maximum use.



Audio equipment tends to be ratherfashion conscious and so the cases and cabinets used for amplifiers and tuners change styles rapidly. Because this type of project is on show to anyone who enters your home it is very important to ensure that the workmanship put into the cabinet and front panel of any audio project is of the very best. Probably the easiest way to achieve this is to build from a complete kit where the cabinet is provided. However, if you are designing your own circuits or insist on making the entire unit yourself then there are a few recent additions to several makers' case ranges which would lend themselves to audio equipment.

For the amateur radio enthusiast the need is for a functional but still attractive case at a respectable price and it also probably needs to be all metal as well.

Metalwork

Although very respectable work can be turned out on the kitchen table using nothing more than a pair of scissors, an old file and a simple hand drill if you have the patience, it is very much easier with a few basic tools.

The home constructor needs to be able to drill holes in metal or plastic panels, file rectangular holes for meters or plugs, cut pieces of metal to size and shape and bend them into simple shapes such as brackets. To perform these operations a few basic tools are essential and you must learn how to use them properly.

Some means of holding the work is essential if painted or polished surfaces are not to be scratched.

Ensure that your cutting tools and drills are all kept sharp and in tip-top condition. Blunt and rusty tools do not

Practical Wireless, March 1979

Drilling sizes

Thread	Clearance	Tapping
OBA 2BA 4BA 5BA 6BA 8BA 10BA 12BA	6.10mm 4.80mm 3.70mm 3.30mm 2.85mm 2.25mm 1.80mm 1.40mm	5.10mm 4.00mm 3.00mm 2.65mm 2.30mm 1.80mm 1.40mm 1.05mm

A professional look can be achieved with the *PW* Front Panel Overlay System. This picture shows the overlay for the *PW* Purbeck oscilloscope. The overlay can be used over a plain white card panel or over a coloured one if preferred. A thin Perspex sheet can be fitted over the film to protect it and hold it firmly in place

help to make a good job. Mark out the positions of the holes and cut-outs accurately, paying particular attention to ensuring that groups of holes are correctly positioned relative to each other. Centre-pop each hole before drilling a small diameter pilot hole first then following up with the correct size drill. Rectangular cut-outs are made by drilling holes at each corner and using an Abrafile fitted in the hacksaw frame cut along the four sides. Use a suitable file to finish the cut-out to size.

Wherever possible try to use components that have some sort of bezel to cover up the holes. This gives you much more leeway with your metalwork.



Collet type knobs provide a firm fastening together with ease of positioning on the shaft. These are Sifam knobs and the various component parts can be seen



Front Panels

Lettering on the front panel can make or mar a project. Although press down letters and numerals can be used with good effect they are not as easy to use properly as is widely imagined and unless they are carefully fixed with special varnishes they tend to rub off quickly in use.

A system to enable constructors to produce professional looking hard wearing front panels for many *PW* projects has been evolved.

This uses a photographically produced transparent film of the front panel which can be carefully cut out and placed over the main panel. A thin sheet of Perspex can be placed over the film to hold it flat, the "sandwich" being held in place by the controls.

A further stage in the production of front panels is to use the film overlay as a photographic master to make a thin metal panel which is then stuck onto the main front panel. This method produces superb panels but does require the use of ultraviolet lamps, specially sensitised metal sheet and the appropriate chemicals.

Knobs

To complement the front panel design you should choose suitable knobs, bearing in mind the use to which the unit is to be put. For test equipment where ease of use is vital it is difficult to better a collet type of knob. These fit tightly onto the control shaft by a simple collet action and do not require a flat on the shaft for a grub screw.

For audio units more fashionable knobs can be chosen but remember that good knobs cost more than poor ones.

Wire sizes

s.w.g.	dia. mm	s.w.g.	dia. mm
10	3.251	30	0.315
12	2.642	32	0.274
14	2.032	34	0.234
16	1.626	36	0.193
18	1.219	38	0.152
20	0.914	40	0.122
22	0.711	42	0.102
24	0.559	44	0.081
26	0.457	46	0.061
28	0.376	48	0.041

Basic tool kit

Hand drill with $\frac{3}{8}$ inch chuck Drills 1mm, $\frac{1}{16}$, $\frac{1}{8}$, $\frac{1}{4}$, $\frac{3}{8}$ inch diameter, countersink bit Junior hacksaw Abrafiles (to fit hacksaw) Centrepunch Small ballpein hammer Small files including round, halfround and three square Steel rule, 12 inches long Small engineer's square Scalpel and blades (also use as scriber for marking out) Small vice Screwdrivers Pliers, small and large Sidecutters **BA** spanners Soldering irons 15W, 30W

The mystic art of soldering

Assemble your projects correctly

The absolute success of any electronic or wireless project lies with the workmanship put into the construction. This applies especially to the quality of the soldering. One dry joint and the entire project is likely to fail to work, causing despair to the beginner.

It is simply not sufficient to assume that a project can be thrown together anyhow, the solder blobbed onto the joints, the power switched on and it will work. With modern circuits and exotic components this is just not the case. The would be wireless constructor should aim to master the various constructional techniques used.

The basic art which must be perfected before starting any constructional project is the technique of soldering.



A soldered joint is used for two purposes in electronics. The first is to provide a mechanical fastening for the various electronic components.

The second and main purpose of the solder is to provide a low resistance electrical path between the leads of the components. It can only do this if the solder "wets" each lead to be connected.

The art of good soldering is not difficult to master and requires the joint to be clean and free from greases, the iron to be really hot and perfectly clean and the correct grade of solder to be used.



Choice of Iron

Let us discuss the choice of iron first as this has a direct bearing on how easily you can make a given joint. The only purpose of the iron is to heat the metals to be soldered to the correct temperature so that the solder flows properly. It is not intended to be used as a means of carrying the solder to the joint or as a tool for bending lead ends over. The iron should ideally be matched to the joint being made so that the temperature is neither too hot nor too cold. If the temperature is too hot it is probable that the more delicate components might be damaged while if it is too cold the solder will not melt properly and the joint will be dry.

Soldering irons are available in a wide variety of sizes and shapes. The first figure to be quoted when talking about irons is the wattage of the element and these vary from about 10W up to about 240W—ideal for soldering the seams on your car radiator. For our purposes a 15/17W miniature iron with an iron plated copper bit of about 2.3mm diameter is ideal for modern p.c.b. work

Top left is a badly made soldered joint on a p.c.b. probably made with an iron that was cold and dirty. Too much solder has been used and has bridged across the three pads. Below it is the same area of the board after the excess solder had been removed and the joint made with a clean hot iron. Top right shows a solder bridge across two adjacent pads. Below are the same joints properly made. The heading picture shows a Stiron 75W iron and Antex 30W and 17W miniature irons and will cope with most joints to be found on *PW* project boards.

However where large diameter earth wires, metal screens and components with heavy gauge wires are met with then such a small iron will not be capable of supplying enough heat to bring the metals up to soldering temperature rapidly enough and a poor joint will be the result. For this type of work and also for older projects such as those using valves a 25/30W iron with a bit of around 3.2mm diameter is needed.

If you want to go in for some really heavy brass metal chassis bashing then you will need at least a 60W iron with a large bit and you might even find that a small propane gas blowtorch is better.

Preparation

When preparing the components for soldering ensure that the leads are clean and, if necessary, tin them by applying the hot iron to the leads together with a resin cored solder so that the lead is completely covered by a thin layer of solder.

The leads should then be carefully bent so that the component fits into place without any strain on its leads. Place the hot iron onto the joint and immediately apply resin cored solder to the joint. The solder should flow easily and quickly to completely cover the joint, when the solder and iron can be removed and the joint allowed to cool. It is very important that the components are not allowed to move during the cooling down period as this will affect the quality of the joint.

When it has cooled down and the solder solidified the joint can be inspected. The solder should completely surround the joint and be bright and shiny. If the joint is dull or crystalline in appearance the joint is "dry" and must be remade. Likewise any small blowholes or areas where the solder has not "wetted" the leads means that the joint is suspect.

One very common fault is to use far too much solder on each joint. Only enough solder should be applied to the joint to completely cover it. An excess of solder is not only untidy, and could cover a bad joint but it is also expensive.

If you find that you have to remove a component from a p.c.b. at any time this can be easily done by using a de-soldering



These two pictures show a TV game p.c.b. as built by a novice (above) and after rebuilding correctly (below). Note how the components are poorly formed and placed in the upper picture with the transformer hanging on one small screw. All components should be neatly formed and positioned with great care and precision to obtain maximum reliability



braid. This braid is placed on the joint and a clean iron applied. The solder melts and is literally sucked into the braid by capilliary action. A clean piece is used for each joint, of course.

The soldering iron bit must be kept clean and this is best achieved by using a damp sponge and wiping the hot bit with it after each joint is made. Several stands are available to hold the iron when it is not in use and these usually incorporate the sponge in their bases. Remember to keep the sponge damp though.

Never use any other solder than a resin cored variety which has been specially formulated for electrical and electronic work. Fluxes such as Baker's Fluid or other highly active varieties must never be allowed to get anywhere near an electronic component, the results will definitely prove disastrous for the component as well as shortening the life of your soldering iron bit. Leave these corrosive fluxes to the plumbers.

Identifying components

The resistor and capacitor colour code explained

Although there are some minor differences in specific capacitor ratings, both resistors and capacitors follow the same colour code. In the case of resistors, where the colour code is in use, it refers to a nominal value in ohms, the unit of resistance. There are of course other types of resistor which do not utilise a colour code, such as wirewound types (high current or high stability) and some which simply have the resistance value written as a figure on the resistor body itself, but the vast majority of carbon, metal oxide, and thick film types use the colour code.

The way in which the code works is very simple, the colours being read off from one end of the resistor to the other, beginning at the end where the colours are concentrated. The first ring of colour indicates the first digit, the second ring the second digit, and the third indicates the multiplier or number of zeros in use. The fourth colour indicates the tolerance over the stated range, brown indicating 1%, red 2%, gold 5%, and silver 10%.

A typical resistor might read yellow, purple (or violet if you prefer), red, and gold. This indicates 4700ohms, variously written as 4.7k, 4.7k Ω , or 4k7, and a tolerance of 5% over that range. It is important to appreciate that the third colour (the multiplier) actually denotes the number of zeros, thus a 47 Ω resistor would appear as yellow, purple, black, with a tolerance band following, indicating that there are *no* zeros in the multiplier.

In general, the colour code is restricted to carbon or metal oxide types, and these resistors will of course be suitable for all types of circuit application. On the other hand, wirewound types, which normally have their value stamped or printed on the body of the resistor, will possess inductive properties, which may render them unsuitable for r.f. circuits.

The stability and tolerance of a particular resistor is to a great extent dictated by the materials of which it is made, as is the inherent electrical noise which it produces. High stability items are typically constructed from carbon film (5% tolerance) or metal oxide (2% tolerance) cermet ("thick film") also 2% tolerance, but precision items are of course wirewound, giving a rated tolerance of 0.1%.

In the case of power dissipation, wirewound forms are inevitably the favoured construction, giving a dissipation in specific items up to about 50W commercially, but once again, fairly high levels of dissipation can be managed with moulded carbon compound types, up to about 2W.

Where low noise levels are concerned, metal oxide and thick film types are used.

Capacitor Markings

Although the colour code is the same as that for resistors, it is generally



the second s			
COLOUR	1st & 2nd BAND	MULTIPLYING FACTOR	TOLERANCE
BLACK	0	1	_
BROWN	1	10	+1%
RED	2	102	+2%
ORANGE	3	103	
YELLOW	· 4	104	_ 1
GREEN	5	105	_
BLUE	6	106	_
MAUVE	7	107	- 1
GREY	8	108	_
WHITE	9	109	_
SILVER	-	10-1	± 5%
GOLD	-	10-2	±10%

Table of Multipliers

(referring to all electrical			
functions—current, voltage,			
resistance, inductance, fre-			
quency, power, and time)			
$p = pico = \times 10^{-12}$			
n = nano = ×10 ⁻⁹			
$\mu = micro = \times 10^{-6}$			
$m = milli = \times 10^{-3}$			
$k = kilo = \times 10^3$			
$M = mega = \times 10^{6}$			
$G = giga = \times 10^9$			

restricted to ceramic disc, tantalum bead, and some moulded or dipped types. In the case of capacitors, there is some extra information necessary, notably the working voltage. The general code is read in the same way as with resistors, beginning at one end and moving towards the wire ends, the last line of colour indicating the working voltage. In the most common form, the polyester dipped type, red = 250V and yellow 400V. In these types a white tolerance band indicates 10% and black 20%.

Caution must be exercised where colour bands are similar to the colour of the capacitor body, as for example, a 2200pF (2·2nF or 2n2) polyester dipped type might well appear as a totally red item, made up from red = 2 (first digit), red = 2 (second digit or tens), and red = 2 (multiplier or number of zeros).

Tantalums are usually low voltage types and the code, again, is read from the free end, away from the wire ends. The first colour is the first digit, the second the tens in the figure, the next the multiplier, and the final colour the working voltage. These voltage bands follow a colour code, black indicating 10V, yellow $6 \cdot 3V$, green 16V, blue 20V, grey 25V, white 30V and salmon pink 35V.

Applications

The non-electrolytic categories, such as silvered mica, ceramic, paper, polyester and polycarbonate are highly suitable in all signal processing applications, both audio and radio, with the principal restriction being each type's working voltage and temperature capabilities. Ceramics are normally used in situations where their good temperature characteristics are an advantage, such as tuned circuits. Polyester, paper and polycarbonates are universally useful in high voltage situations where good insulation resistance is required such as pulse circuitry in television, and general signal and power stages in radio and audio circuits. In the main, coupling in transistor a.f. stages etc. utilises specific polvester items, due to their smaller physical size and good tolerance.

Electrolytics are used where high capacitance values are needed, and in the de-coupling of power supplies, where hum and ripple may be a problem.

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Component buying

How and where to purchase those elusive parts

"Where can I buy the components for the *PW* 'Whathaveyou' project?" is one of the commonest queries received. This article will try to explain the pitfalls of purchasing electronic components, and show how to overcome them.

Unless you can buy a complete kit for a project it is quite probable that you will have to undertake some detective work to uncover the relevant suppliers.

The first thing that any serious home constructor needs is a library of catalogues from various suppliers of components. A run through the adverts in *PW* will indicate who produces a catalogue and you really cannot have too many of them. Using your library you can select the appropriate supplier for most components. You can compare prices for various components to try to obtain the best deal if this sort of exercise takes your fancy.

Exotic projects

Consulting your favourite supplier's catalogue is all right for the common or garden components such as resistors, capacitors and semiconductors, but what about the special items often specified for the more exotic projects?

Often these are available only from one source and where this is so the source is given in the components list or in the text of the article describing the project.

One name that keeps appearing in component lists is that of RS Components. This company is a component distributor who produce a comprehensive catalogue widely used in industry. All the components sold by RS are "own brand" and they only sell to *bona fide* trade customers. Because of their rapid service and well-produced technical catalogue, their components are used widely by authors when preparing projects, and the staff of PW also use them for similar reasons.

This leads to complications when a components list calls for an RS Components part number as many readers will find it difficult to obtain them. One way out is to befriend your local radio and TV repair shop who will certainly have an account with RS. It must be remembered that the prices given in their catalogue are not retail prices, and do not include VAT either, so be prepared to have to pay around 50% above catalogue price.

Up until recently RS Components' sister company Doram supplied to the amateur market but this service has been withdrawn. However several of the regular advertisers in *PW* will obtain any RS Components part on request.

By the way, try and persuade your friendly repairman to let you have an old issue of RS Components' catalogue, it is a mine of valuable information on components, giving such information as sizes, connections and specifications.

Cases and cabinets

When it comes to the cabinet or case to fit the project, these are usually available from several advertisers, most of whom carry a range from two or three manufacturers. In most projects the actual case style is a matter of personal choice and the one used by the author need not be copied. However, some projects are built using specially designed cases or utilise some particular feature of a case, and then it is important that only that case is used.

If you are the type who enjoys making your own chassis and cases then you will need to find a supplier of raw materials. This is not always easy and it pays in this case to have a copy of *Whiston's Catalogue* (K. R. Whiston Ltd., New Mills, Stockport SK12 4PT). This lists small quantity sizes of aluminium, brass, steel, etc., and is available for an s.a.e.

Printed circuit boards for PW projects should be obtainable from advertisers. but, individual readers can, if they have the equipment, make their own boards from the copper track patterns given fullsize in the magazine. Another way is to take the track pattern to your local printed circuit board company and get them to make you a board. This is likely to be expensive as they will have to make you a photographic master from the drawing first (note that this must only be made for you and boards must not be made for resale by this method without obtaining permission and paying royalties to the copyright holder). You can find the address of p.c.b. makers in your local "Yellow Pages", which can also prove useful in locating component stockists, metal stockists, etc.

Semiconductors

When buying semiconductors for a project you will have to make up your own mind as to whether you risk using cheap unmarked types or spend out more money for guaranteed ones. Most suppliers will honour the maker's warranty and change faulty i.c.s or transistors so long as they have not been soldered into a circuit. With c.m.o.s. types, do not be tempted to take them out of their packaging to look at them. They are very prone to damage from static which builds up on all humans these days. Leave these i.c.s firmly in their conducting foam or silver foil wrappers until you are ready to insert them into their sockets. If you are supplied with c.m.o.s. not protectively wrapped in foil, conducting foam or a special housing, get in touch with the supplier immediately. If possible it is best to buy such components only from suppliers who are recognised or franchised by the manufacturer.

Finally, if you have exhausted all the above possibilities and feel that you simply must ask the magazine staff, please enclose a s.a.e. and don't expect miracles, and please, on please, only after you really have tried yourself!

Handling c.m.o.s.

Avoid killing your i.c.s with static

Unlike t.t.l. (transistor-transistor logic), c.m.o.s. (which stands for "complementary metal oxide semiconductor") devices are very prone to damage if handled or fitted into circuits without due precautions having been taken.

It is true that one or two digital c.m.o.s. devices are fairly robust, and can usually only be destroyed by reversing polarity of the supplies or by feeding a high signal at low impedance into the inputs while the device is without a supply, but general c.m.o.s. "chips" for use in the audio, video, music and radio fields are highly susceptible to static charges and extremes of temperature.

A great deal has already been written on somewhat amusing lines concerning the wearing of nylon clothing while working on such devices, but the basic facts are quite simple and need no "scripting" in order to emphasise the requirements.

- Unused inputs should always be tied to ground or positive supply, depending upon requirements.
- Polarity of supplies must be carefully checked before any connections are made.



As it should be, neat and tidy.

- 3) Low impedance sources, including charged electrolytic capacitors, must not be connected directly to the input terminals, especially in the case of logic "chips". It is essential to use a surge limiting resistor in such cases of at least 1000Ω .
- Input terminals must not be allowed to "float" and must be, like unused inputs, tied to ground or positive via a high resistance.
- 5) Where devices are supplied in nackages of conductive foam, conducting foil, or the specialised conducting tubes which are becoming typical of c.m.o.s. packaging, they should not be removed until the very last moment in order to avoid the effects of any local static charges.
- 6) It is as well to avoid wearing nylon shirts or similar synthetic



fibre clothing while working on these "chips", and the same point applies to plastic coverings for bench tops, which should be removed.

- Where possible, the device should be inserted directly into the i.c. holder without any actual contact being made between operator and device.
- In extreme cases, the body can be earthed via a loose chain on the wrist, this technique being used with operatives during manufacture. A metal-covered bench is also used.
- 9) The pins of the device should not be touched with the fingers.
- 10) The device must not be disconnected or otherwise removed from the associated circuitry while it is functioning.
- 11) When soldering to such devices, an earthed bit must be used.

Care taken in handling, or not handling, c.m.o.s. components will greatly increase the reliability of projects which use this type of integrated circuit. Properly used c.m.o.s. offers a lot of advantages for the hobbyist and attention to the advice given above will ensure your confidence in the finished project.

How not to build your project. This untidy "heap" shows many faults such as trying to fit an 8-pin d.i.l. package i.c. into a p.c.b. designed for a TO99 circular package by tacking short lengths of wire to each lead to extend it

Practical Wireless, March 1979