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## QUERIES

While we will always try to assist readers in difficulties with a Practical Wireless project, we cannot offer advice on modifications to our designs, nor on commercial radio, TV or electronic equipment. Please address your letters to the Editor, Practical Wireless, at the above address, giving a clear description of the problem and enclosing a stamped self-addressed envelope. Only one project per letter please.
Components are usually available from advertisers. A source will be suggested for difficult items.

## SUBSCRIPTIONS

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## NEWS \& VIEWS

## Production Lines

Information on the latest products

## FOR OUR CONSTRUCTORS

25 PW ''Soundlite"' . . . . . . . . . . . T. P. Hopkins An economical 3-channel sound-to-light unit
Ideas Department
Step Tone Generator. Car Cassette Power Supply
PW "Hythe" Marine Band Receiver-2 . M. Tooley \& D. Whitfield Using the receiver, plus some add-on units
PW '"Winton'" Stereo Amplifier-1 . . . . . . E. A. Rule A high quality, 50W per channel design, using power f.e.t.s
Follow-up to the PW "Avon" . . . . . . Peter Preston Increased power and 12 V operation
Wide-band Noise Source
. D. Whitfield
An r.f. noise generator for receiver testing
Tone-burst Generator
P. Hodson

An audio oscillator for v.h.f./u.h.f. repeater access

## GENERAL INTEREST

IC of the Month
Brian Dance
The TDE1607 interface device
On the Air
Amateur Bands . . . . . . . . . . Eric Dowdeswell
MW Broadcast Bands . . . . . . . . . Charles Molloy
SW Broadcast Bands . . . . . . . . . Charles Molloy
VHF Bands Ron Ham
VHF Personalities-The Two Erns . . . . . . . Ron Ham

## $\approx$ FREE THIS MONTH

"HINTS \& TIPS FOR CONSTRUCTORS" - A special 8-page
supplement on all aspects of construction

We regret that, owing to circumstances beyond our control, the size of this issue of Practical Wireless has had to be reduced at short notice by eight pages. We apologise to readers for any disappointment caused.

## B. BAMBER ELECTRONICS

Dept. P.W. 5 STATION ROAD, LITTLEPORT, CAMBS, CB6 $10 E$ Telephone: ELY (0353) 860185 (2 lines) Tuesday to Saturday

|  | PLEASE ADD 8\% VAT UNLESS OTHERWISE STATE |  | TCP2 TEMPERATURE CONTROLLED IRON. <br> Temperature controlied iron and PSU. $\mathbf{£ 3 0}$ VAT (E2.40). <br> SPARE TIPS <br> Type CC single flat. Type $K$ double flat fine tip Type P. very fine tip $\mathbf{£ 1 . 5 0}$ each + VAT (8p) MOST SPARES AVAILABLE. |
| :---: | :---: | :---: | :---: |
| coil) to fit $\frac{1}{2}^{\prime \prime}$ reeds (not supplied) 2 for 50p. | CELESTION $8^{\prime \prime} \times 5^{\prime \prime}$ ELIPTICAL | SPECIAL DFFER FOR CDMPUTER BUILDERS, ETC. <br> 19 way ritbbon cable, decimal coded 4 metres for 1.25 . 13 way heavy duly ribtoin cable, decimal corted, lideal for PSU runs) 3 matres for $£ 1.50$. |  |
| fe | ohm, 3 watts |  |  |
|  |  |  |  |
| DUAL TO18 HEATSINKS $1^{\prime \prime} \times \frac{1}{2}$ " $\times \frac{1}{4}$ " with screw-in clamps. 3 for 50p. | $\times 1^{\prime \prime}$ high, with integral heatsink, com | CLAREED REED RELAYS, complete with reeds. TYPE 1. Size approx. $2 \frac{1}{2}^{\prime \prime} \times \frac{3^{\prime \prime}}{4} \times \frac{1^{\prime \prime}}{2^{\prime \prime}} 1$ pole make, $9 V 400$ ohm coil, 35 peach. TYPE 2 , Size approx. $2 \frac{1}{2} \times 1 \frac{1}{2} \times \frac{1}{2}$ ". 2 pole make +2 pole break. $2 \times 9 \mathrm{~V} 200$ ohm coils, 60 p each. |  |
| AINS TESTER SCREWDRIVERS 100 to V . Standard size $\mathbf{5 0 p}$. Large 70p. |  |  | WELLER SOLDERING IRONS <br> EXPERT. Built-in-spotlight illuminates work Pistol grip with fingertip trigger. High efficiency |
|  |  | VIDICON SCAN COILS itransistor type, but no daral complate with vidicon base $\mathbf{£ 6 - 5 0}$ each Brand New |  |
| SMALL SIDE CUTTERS (with wire holding davice) |  |  | copper soldering tip. <br> EXPERT SOLDER GUN S100D £12,00. EXPERT SOLDER GUN KIT (spare bits, case, etc.) $£ 15-00$. Spare bits 40 p pair. |
| MINIATURE FILE SETS. Set of $6 \mathbf{£ 2} \mathbf{2 0}$. HIGH QUALITY RELAYS, 4 pole C/O, 3A | table batteries from mobite supply). Only needs ane 8 FY50/51/52 or similar transistor. which | IC TEST CLIPS, clip over IC while stilf soldered to pcb or in socket. Gold-plated pins. ideal for experimenters or service engineers. 28 pin OHL $\mathbf{£ 1 . 7 5 , 4 0}$ pin DIL $£ 2 \cdot 00$. Or save by buying one of each for $£ 3-50$. |  |
| contacts. 12 V DC coil. 150 ohm . Size approx. $1^{\prime \prime} \times \frac{3^{\prime \prime}}{} \times 1 \frac{1}{4}$ ", with plastic covers. 80p each or 2 for $\mathbf{f} 1.50$. | ed with a star-type heatsink £2-00 each. |  | IXED COMPONENT PACKS, containing sistors, capacitors, pots, etc. All new. Huneds of items. $\mathbf{£ 2}$ per pack, while stocks last. |
| LARGE ELECTROLYTIC PACKS. Contaí | HE NEW EAGLE |  |  |
| high voltage types. pack $\left[+12 \frac{1}{2} \%\right.$ VAT . | QUEST containing Audio, in-car, and test equipment, etc. | ASS BEAD FEEDTHROUGH I TORS. Solder-in type, overall dia. | BSA AUTOCHARGE RECORD PLAYER DECKS with cue device, 33-45-78 RPM. for $7^{\prime \prime}, 10^{\prime \prime}, 12^{\prime \prime}$ records. Fitted with SC12M Stereo |
| Slider Switches. 2 pole make and break lor can be used as 1 pole change-over by linking the two centre pinsl. 4 for 50p. |  | PLASTIC PROJECT BOXES with screw on lids (in black ABS) with brass inserts. <br> Type NBI approx. $3 \frac{1}{4}^{\prime \prime} \times 2 \frac{1}{4}^{\prime \prime} \times 1 \frac{3^{n}}{4} 45$ fach <br> Type NB2 approx. $4^{\prime \prime} \times 3^{\prime \prime} \times 13^{3 \prime \prime} 55 p$ each <br> Type NB3 approx. $4 \frac{3}{4}{ }^{\prime \prime} \times 33^{\prime \prime} \times 1 \frac{3}{4} \times 65 p$ васh <br> Type NB4 approx. $8 \frac{1_{2}}{}{ }^{n} \times 5 \frac{7}{4} \times 3 \frac{1}{4}^{n}$ E1-50. | Ceramic cartridge and styli. Brand new $£ 14.00$ $+12 \frac{1}{2} \%$ VAT. |
| DUE TO A CHANGE OF SUPPLIER, OUR STOCK ALUMINIUM BOXES AND VINYL COVERED EQUIPMENT CASES WILL BE AS FOLLOWS | ze approx. $3^{*} \times 3^{\prime \prime}$, with large ates. $0-9+$ Clear. A, B, Dual Wate ew only. £2.00 while stocks last. |  | PLAYER DECKS, Model 6.300, with cue device, 33-45-78 r.p.m. for $7^{\prime \prime}, 10^{\prime \prime}$, $12{ }^{\prime \prime}$ records. Fitted with KS41B Stereo Ceramic cartridge and styli Brand new $\mathbf{£ 1 6 . 0 0}+12 \frac{1}{2} \%$ VAT. Please note, record decks sent by Roadline, allow 14 days for delivery. |
|  |  |  |  |
|  | T. |  |  |
| AL1 $3^{\prime \prime} \times 2^{\prime \prime} \times 1^{\prime \prime}$ 60p |  |  |  |
|  | SEMICONDUCTORS | PLUGS AND SOCKETS <br> BNC Plugs, new 50p each. <br> $\mathrm{N} / \mathrm{T}_{\mathrm{yp}}$ Plugs 50 ohm, 60p each. 3 for $\mathbf{E 1} 50$. <br> PL259 Plugs (PTFE) brand new, packed with reducers, 75p each. <br> SO239 Sockets (PTFE), brand new (4-hole tixing typel. 60p each. | FULL RANGE OF BERNARDS/BABAN ELECTRONICS BOOKS IN STOCK. S.AE FOR LIST. |
|  | ult |  |  |
| $6^{\prime \prime}$ | BC108 (metal can), 4 for 50p. |  |  |
|  |  |  |  |
| AL6 $8^{\prime \prime} \times 6^{\prime \prime} \times 2^{\prime \prime}$ |  |  | VARICAP TUNERS Mullard type ELC1043/05. Brand New. $\mathbf{£ 5 \cdot 0 0}+12 \frac{1}{2} \%$ VAT. |
| AL7 $\mathrm{Cl}^{\prime \prime} \times 6^{\prime \prime} \times \mathbf{3}^{\prime \prime}$ ¢1.75 | BCY 72 Transistors, 4 for 50p. 12 for 25 p. |  |  |
| Viny\# Equipment CasesBlte Viny covered steel tops with plain | E, TOS |  | BARGAIN PACK OF LOW VOLTAGE |
|  | BF1 52 (UHF amp/mixer), 2N3819 Fet. 3 for 60p. | SOLDER SUCKERS (Plunger type). Standard Model, $\mathbf{E 5 \cdot 5 0}$. Skirted Model. E6. Spare Nozzles 60p each. |  |
| BCO |  |  | ELECTROLYTIC CAPACITORS. Up to 50 V working. Seatronic Manufacture. Approx 100 $\mathbf{E 1} \cdot \mathbf{5 0}$ per pack $+12 \frac{1}{2} \%$ VAT. |
|  |  |  |  |
| BC2 $6^{\prime \prime} \times 4^{\prime \prime} \times 3 \frac{10}{}{ }^{\prime \prime}$ ¢2.25 |  |  | Dubiller Electrolytics, 50 F. $450 \mathrm{~V}, 2$ for 50p. Dubiller Electrolytics, $100 \mu \mathrm{~F} .275 \mathrm{~V}, 2$ for 50p. Plessey Electrolytics, $470 \mu \mathrm{~F} .63 \mathrm{~V}, 3$ for 50p. |
|  |  | NEW MARKSMAN RANGE OF SOLDERING IRONS. |  |
| BC4 $10^{\prime \prime} \times 6 \frac{10}{\prime \prime \prime} \times 3^{\prime \prime}$ |  |  |  |
| BC7 $\quad 12^{n} \times 6 \frac{1}{2}^{\prime \prime} \times 5^{\prime \prime} \quad$ £3.25 |  | ING IRONS. |  |
| MAINS TRANSFORMERS. Type $15 / 300$ <br> 240 V  <br> input. 15 V at  <br> 300 mA output.  <br> $\mathbf{£ 1 . 5 0}$  | 15 A .90 Watts, Flat pack type. 2 for $£ 1-50$. | S125DK 25W 240V + bits etc., KIT £5.50. BENCH STAND with spring and sponge for |  |
|  | GERMANIUM DIOOES, approx 30 for 30p. |  | Dubiller Electrolytics, 5000 F, 50V, 60p each. ITT Electrolytics, 6800 F, 25V, high grade |
| each <br> MAINS TRANSFORMERS. Type 45/100, 240. 220, 110 , OV input. 45 V output. $£ 1.50$ each. | 741 CG op amps by RCA, 4 for $£$ | Spare bits MT9 (for 15 W ) 60p, MT5 (for 25 W ) 50p, MT 10 (for 40W) 55p. <br> ALL PRICES + 8\% VAT. | screw terminals, with mounting clips, 50p each. PLEASE ADD $12 \frac{1}{2} \%$ VAT TO ALL CAPACITORS. |
|  | RED. LEDs (Min (YPe) 5 for 70p. |  |  |
|  |  |  |  |
| pase: ' | DDER, MIMTMUW ORDER E2. ALL ALL GOODS IW STOCK DESPATCHED BY | ICES IWCLODE POST \& PACKIWG, RW. CALLERS WELCOME BY APPOWYTL | OWLYI, SAE with ALL ENOUIRI |

## OPEN UP THE EXCITING WORLD OF SHORT WAVE LISTENING


spx 30
For the advanced, keen short wave listener the choice of receiver has usually been between cheap and nasty or very good but very expensive equipment. We think that the SRX- $\mathbf{3 0}$ will provide that listener with excellent performance at a reasonable cost and is the answer to this eternal problem. The SRX-30 is based on an advanced drift cancelling loop system which gives spot on dial accurac at any frequency between 500 KHz and 30 MHz together with easy to understand frequency readout Suitable for all users from raw beginners, thanks to it's simplicity of operation: to experienced listeners and amateur operators thanks to it's advenced technology, the SRX-30 is the best communication receiver available in its price range today.
Completely self contained, including operation from mains or 12 volts de. the SRX-30 is at home on broadcast or amateur bands. All mode reception oi AM. CW. LSB. and LSB is provided and receive Sand for futl details today or give us a ring and we'll tell you all about the SPXCarriage by Seuaricor $£ 3$.


NEW. Bellsonic power supply giving fully regulated 12 V dc output at 3 amps continuous rating rating from $220-240$ V ac mains input.
uses for the amateur experimenter or professional user. Incredibly low price £17.28

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We also stock the RAK Listener 3 aerial systen for the man who demands the best and has the room for it. Double dipole system complete sith H/D alloy wire, insulators, coaxial feeder. centre connector etc. in an ov
$\mathbf{P} \& \mathbf{R}^{66 p}$.
Also available is the Listener 1 loaded end fed single wire aerial system for restricted space situations, Overall length only is feet due to use of loading coil and element folding. Ideal aerial for any user of short wave bands. Price $£ 7.50 \mathrm{inc}$, val. P \& P86p.
If you need professional osciltoscopes at reasonable prices, please contact us for details of terrific
TRIO range. Full range of other test equipment stocked.
For all that's good in Amateur Radio, contact
LOWE ELECTRONICS LTD., 119 Cavendish Road, Matlock, Derbyshire. Tel: 06292430 or 2817. For full catalogue, simply send 45 p in stamps and request catalogue CPW.

NEW. CL22 aerial tuner which will match New. CL22 aerial tuner which wil match frequency between 1.5 and 30 MHz . Six switch-
ed ranges with fully tunable receiver and aerial ed ranges with fully tunable receiver and aerial marching capacitors. A worthwhile addition to any SWL station and an instant improvement in aerial matching problems. Price $£ 15.75$ inc. vat
$\mathrm{P} \& \mathrm{P} 66 \mathrm{p}$.

# FABULOUS PROFESSIONAL DISCO SYSTEM 

F.A.L. De Luxe PROFESSIONAL Carr. GONSOLE (Powered)
etc. £15

Slide Fade Controls. Autofade with Music Overide on both mic. and jingle inpuis. Headphone pre-fade monitor with 5 Push-button selector. Illuminated V.U. meters. Illuminated unity gain. Mic. channel with Bass \& Treble. Slave outlet. Treble Control (Music Channel) Output into line (Slave) Treble Control (Mic. Channel) Autofade Recovery Bass Control (Music Channel) Microphone Input. Bass Control (Mic. Channel) Tape/Jingle. 1/P


## PAIR MATCHING FULL RANGE De Luxe 80w LOUDSPEAKERS

Each inc. Pair of Powerful $12^{\prime \prime}$ Bass units (with aluminium centre domes), and High Frequency Horn unit to extend frequency range to above 17 kHz . Normally 889.95 ea.

## FANTASTIC SPEAKER OFFER

TWIN 12" SPEAKER CABINET PLUS PAIR 12" SPEAKERS of Robust vibration-proof construction Fitted protective corner pieces, Re-
movable Vynair covered front with silver effect trim. Sunken jack socket with escutcheon at the rear Pair $12^{\prime \prime} 15 w$ speakers for wiring up and front mounting While stocks last in above supplied to Three items complete a 30w unit for
P.A.

£19.95


and
Turntables BSR P200 Beltdrive are Garrard 125SB Belt-drive with low and magnetic cartridge per system $£ 10$ for Garrard only.
POWER: POWER: 120 watts continuous RMS into 4 ohms able flexi-beams over each turn table with independent switching. CABINET: Solidly made to withstand the rigours of transporting. Covered in heavy duty leather cloth type material in attractive colour combinations Corner cap protectors Recesed ararying harates
List £320.76


ALL THREE UNITS PRODUCED BY BRITAINS LARGEST MAKERS OF DISCO EQUIPMENT


TRANSMITTER RECEIVER. MK.123. Very compact Army unit for use in range 2.5 to $20 \mathrm{Mc} / \mathrm{s}$, receiver section 7 valves inc RF stage \& BFO provides O/P for 4 K ohm phones, 3 bands with direct cal 2.5 to 5,5 to 10 \& 10 to $20 \mathrm{Mc} / \mathrm{s}$. Tx section 2.5 to $20 \mathrm{Mc} / \mathrm{s}$ in 3 ranges $\mathrm{O} / \mathrm{P}$ 15/25 watts over range C.W. only, 2 valves Crystal Osc \& 5B251M P.A. stage, will match into the following Aerial loads 25, 100,500 \& 1500 ohms as int tune up meter, reqs crystals type FT243 in range 2.5 to $10 \mathrm{Mc} / \mathrm{s}$ int morse key fitted with plug for ext key. Mains P.U. self contained unit for 115 or 200/250v AC overhaul size inc Rx, Tx \& P.U. $30 \times 9 \times 14 \mathrm{Cm}$ weight 4 Kg . also supplied with ext invertor unit for 12 V DC. Supplied in clean cond with 80 page handbook giving circ details etc (no details on 12 v P.U.) $£ 54$.

TAPE RECORDERS. Made for use in language lab equipment 240 v I/P uses BSR type TD. 103 speed deck $5^{\prime \prime}$ spools, two chan transis amps with provision for playback \& record can be used for stereo. P.U. and circ boards are mounted below tape deck approx size $12 \times 11 \times 7^{\prime \prime}$ intended to work phones. Supplied in clean condition may be less knobs \& ind lamps some circ details supplied. No ext case. $£ 13$ also sim valve unit TD. 2 deck $£ 8.50$.
U.H.F. CAVITIES. New spares for 2C39/7289 valves will tune over range $990 / 1040 \mathrm{Mc} / \mathrm{s}$ with int fittings circ supplied $£ 6 \cdot 50$, also Rx section tunable preselector 1080/1130 Mc/s 4 section with 1 N 21 mixer diode for $60 \mathrm{Mc} / \mathrm{s}$ IF with circ $£ 4-50$.
ELECTROSTATIC VOLTMETERS range 0 to 15 KV AC or DC usable scale 3 to 15 Kv complete in wood carrying case $8 \times 9 \times 6^{\prime \prime}$ f10.80.
HELIPOT DIAL standard 10 turn type to fit $\frac{1}{4}^{\prime \prime}$ shaft size $1 \frac{3}{4}{ }^{\prime \prime}$ dia. f 1.50 or with 100 K helipot $£ 2$.
RECTIFIER UNITS ex-Army unit 200/250v I/P provides two DC $\mathrm{O} / \mathrm{Ps}$ of 12 vDC at 3 amps ea can be connected for 12 v 6 amp or 24 v 3 amp will do 4 amp okay for battery chargers good cond f 10.80 .
INFRA RED LAMPS light \& heat units sealed beam 115 v 500 watts size $7^{\prime \prime}$ dia $4 \frac{1}{2}{ }^{\prime \prime}$ deep new American G.E. okay for paint drying etc two for $£ 5.40$.
V.H.F. TEST SET provides RF O/P over range 20 to $88 \mathrm{Mc} / \mathrm{s}$ in 4 bands, as int $2 \mathrm{Mc} / \mathrm{s}$ check, noise generator with 50 Ma meter, int pulse or C.W. O/P complete in case with cal charts \& Ae rods, circ \& notes. These req ext supplies of $250 \mathrm{v} \& 6 \cdot 3 £ 13$.
CRYSTAL OVENS. Miniature type to take $1 \mathrm{Hc} 6 / \mathrm{u}$ or 2 Hc 18 crystals $12 / 24 \mathrm{v}$ operation size inc base $2 \times 1 \frac{1}{4} \times \frac{3}{4}{ }^{\prime \prime}$ new $£ 1 \cdot 20$ either type.
PANEL METERS. 100 Ua scale 0 to 100 linear $2^{\prime \prime}$ dia $£ 3$ also 1 Ma FSD special scale $2^{\prime \prime}$ dia $£ 1.30$ both new.
FREQ METERS type BC 221125 Kc to $20 \mathrm{Mc} / \mathrm{s}$ with int crystal check, with charts \& book require 135 v HT \& $6.3 \mathrm{v} £ 27$.
HEAD \& MIKE SETS for use with 19 set m.c. type with hand mike nom 100 ohm new £4.
HANDSETS No. 3 m.c. with press to take swt suitable 19/62 sets elec okay store soiled $£ 2.50$.
HEATERS flat tubular 240 v 150 watt approx size $6 \frac{1}{2} \times 10^{\prime \prime}$ new £1.80.

DYNAMOTOR UNITS 24 v DC I/P O/P 200 \& 400 v DC 280 Ma int, by removing fan these can be used as powerful $12 / 24 \mathrm{v}$ DC motor with shaft $1 \times \frac{t^{\prime \prime}}{4} £ 6.50$.

CABLE miniature 25 core flex non-screened $5 / 16$ ths osd colour coded new 10 Mts for $£ 3$. Single flex 23/006 okay for Ae wire new 25 Mts for 85 p.
CRYSTALS mixed 10 X \& 10 XJ types in range 5 to $8.5 \mathrm{Mc} / \mathrm{s} 20$ for £2-20.
I.F. TRANS min type 465 Kc new with sec tap 3 for $£ 1 \cdot 30$.

COAX PLUGS \& SKS standard type ex new equip 5prs for £1.30. THERMISTORS bead type 3.8 mm dia 160 ohm at $200^{\prime} \mathrm{c} 20 \mathrm{k}$ at 20'C new 80p
RADIO STATION KIT contains, long wire aerial, dipole aerial, 75 ohm twin feeder, miniature morse key, handset, head \& mike set, misc fittings. In carrying bag new cond. for use in $2 / 10 \mathrm{Mc} / \mathrm{s}$ range. £ 10.80 .
FILTER UNITS 100 Kc crystal filters precision units by Marconi 75 ohm imp in case size $8 \frac{1}{2} \times 3 \frac{3}{4} \times 5^{\prime \prime}$ available in 1 or 2 Kc bandwidths with connectors $£ 5.40$ either type.

Above prices include Carriage \& VAT.
Goods ex equipment unless stated new.
S.A.E. for enquiry or List 21.

## A.H. SUPPLIES

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SEND 15P STAMP FOR THE WORLD'S BEST CATALOGUE OF SPEAKERS, DRIVE UNITS, KITS, CROSSOVERS ETC. AND DISCOUNT PRICE LIST.

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PEERLESS - RADFORD - RAM - RICHARD
ALLAN - SEAS - SHACKMAN - STAG -
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(Dept. P.W.)
SWAN WORKS, BANK SQUARE, WILMSLOW, CHESHIRE SK9 1HF
Discount Hifi Etc. at 5 Swan Street and 10 Swan Street
Speakers, Mail Order \& Export 0625529599 Hi-Fi 0625526213


Low cost version, AOII3 (.02\% dist.) 627.50 (Kit. 623 ). Other instruments include Milivoitmeter, Fachometer, Noise level meter, Distortion Analyser, for lists. VAT extra $8 \%$. Post/Pkg. $£ 1 \cdot 50$.

## TELERADIO ELECTRONICS

325 Fore Street, Edmonton N.9. 01-807 3719
Closed all day Thursday

## Printed Circuit Coils <br> (Pat. App. For)

For the P.W. Sandbanks, only available from the designer of the detector. By reducing the coil capacitance, vast improvements in sensitivity to gold and silver can be achieved and only four resistors need to be changed for maximum sensitivity. A complete kit including the PC coil to fit the Ambit International moulding, the four resistors required, and instructions for only $£ 2.50$ are available from:

PLESSIS ELECTRONICS,
Castle House, Old Road, Leighton Buzzard, Beds.
Callers by appointment only


## ambit <br> international

Production of the new catalogue has been heid up for a few weeks - since we have just been appointed as distributors for two of the most exciting ranges of radio components products yet : The Micrometals range of iron dust torroids cores and formers, and the OKI range of VLSi for digital frequency displays for receivers. We apologize for any inconvenience, but these two ranges are really worth the wait and include some products you will find hard to believe, like the MSM5523 IC, an IC with less than ten external components that gives AM frequency readout to 1 kH from $L W$ to $39.999 \mathrm{MHz}, F M$ frequency readout in 100 kHz steps - (all usual IF offsets programmable by diodes), a 24 hour format clock with 12 hour display, independent on and off timers, time signals on the hours, stopwatch facility and a sleep timer. This costs $£ 14$ with its timebase crystal, and makes all that has gone before an expensive and time wasting excercise. Rather like the way the Intersil ICM7216 has revolutionized the instrument counter market. (See the OSTS ad.) And those of you familiar with Amidon and IG dust torroids, favoured in many new RF designs, will be pleased to know Ambit will be stocking a broad range of the Micrometals types for applications from EMI filters to RF PA stages OK1 frequency counter ICs: details in cat2 MSM5523 for CA LEDs with RHDP such MSM5525 $\begin{array}{ll}\text { as FND507 } \\ \text { for } 31 / 2 \\ \text { digit LCD AM/FM with }\end{array}$ for $31 / 2$ digit LCD AM/FM with
direct segment drive. no clock or timers $£ 11$ inc xtal or timers
Other types for fluore

A brief summary of some of our range of ICs: TDA1062/1.95; TDA1083/1.95; HA1 197/£1.40
CA3123E/£I.40; TBA651/f1.81; CA3089/1.94 HA1137/£2.20; MC1310/£2.20; HA1196/£3.95 KB4424/£2.75; KB4423/£2.53;SD6000/£3.75 KB4412/f2.55; KB4413/f2.75; KB4417/乇2.55 MC1495L/f6.86*; MC1496P/f1.25 LM381N/£1.81: LM1303/\&0.99; ULN2283B/ Other new semiconductor additions: KB4438 milot cancel mox decode HA1370 supercedes TDA2020 TDA1090 HiFi AM/FM TDA 1220 low cost AM/FM new technology in power transistorsis getting heaper. 120 V comp pairs $/ 100 \mathrm{~W}$ for $£ 10.00$ Price reduction on CA3189E ....now $£ 2.20$ New varicaps: to add to the biggest range.. KV1211 2:9v bias to tune MW, like the New pilot tone filters from TOKO...... 208BLR series, individual per channel with a $26 / 38 \mathrm{kHz}$ version for pilot cancel decoder applications. Flat to 15 kHz New crystal filter for amateur NBFM...... ToYO $10 \mathrm{M4B1}$ with over 90 dB adjacent ch .
rejection for 2 m NBFM .10 .7 MHz New ceramir IF fiters for 455 kHz CFM $455 \mathrm{H} 6 \mathrm{kHz} / 6 \mathrm{~dB}$ 15kHz $\mathrm{max} / 60 \mathrm{~dB}$ CFM $455 \mathrm{H} 6 \mathrm{KHz} / 6 \mathrm{~dB}, 15 \mathrm{kHz} \max / 60 \mathrm{~dB}$ -
ideal for MC 3357 etc. sheet of scalas and ledgends for $\mathbf{£ 1 2 . 5}$ Catalogue part 145 p part 250 p all incit Items is generally $121 / 2 \%$, except where marked (*) f3. Phone Brentwood (0277) $216029 / 227050$ 9am-7pm. Callers welcome inc. Saturdays.

## At lust, DIY Hi Fi whith looks us if it isn't.

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$\begin{aligned} & 2 \cdot 9 \mathrm{p} .72329 \mathrm{p} .74115 \mathrm{p} \text {. NE555 23p. } \\ & \text { bc } 182 \mathrm{~b} \text {, bc } 183 \mathrm{~b} \text {, bc } 184 \mathrm{~b} \text {, bc2 } 12 \mathrm{~b} \text {, bc213b. }\end{aligned}$
$\begin{aligned} & \text { bc182b, bc183b, bc } 184 \mathrm{~b}, \mathrm{bc} 2 \uparrow 2 \mathrm{~b}, \mathrm{bc} 213 \mathrm{~b} \text {. } \\ & \mathrm{bc} 214 \mathrm{c} 4.5 \mathrm{p} \text {. plastic equivs bc } 107 \text { bc } 109\end{aligned}$

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| ${ }^{\text {DF96 }} 1.00 *$ | ECF82+ 0.70 ${ }^{\circ}$ | EL84 $\dagger$ | $0.45{ }^{\circ}$ | PC900t | 1.00* |
| DK91 1.05 | ${ }_{\text {ECH35 }}{ }_{\text {ECH42 }} \mathbf{2 . 0 0}{ }^{\circ}$ | EL86 $\dagger$ | $0.75{ }^{\circ}$ | PCC84t | $0.50{ }^{\circ}$ |
| DK92 1.25' <br> DK96 <br> $1.10^{\circ}$ | ECH42 $1.15^{\circ}$ | EL91 | $4.35{ }^{\circ}$ | PCC88 | $0.65{ }^{\circ}$ |
| $\begin{array}{ll}\text { DK96 } & 1.10^{\circ} \\ \text { DL. } & \\ 0.755^{\circ}\end{array}$ | ECH817  <br> ECH83 $\mathbf{0 . 5 5}$ <br> $\mathbf{1}$  <br> 1.25  | EL95 $\mathrm{EL360}$ | 0.80 2.75 | PCC89 ${ }^{\text {P }}$ | $1.05{ }^{\text {c }}$ |
| DL94 1.20 | ECH84† $0.85^{\circ}$ | EM80 | 1.10 | PCC189 | +1.00* |
| OL96 1.10* | ECL801 $0.60^{\circ}$ | EM81 | 1.00 | PCFB2 | 0.96* |
| DY86/7+0.55* | ECL821 $0.60^{*}$ | EM84 | 1.00* |  |  |
| DY802 0.80* | ECL83 $1.50^{\circ}$ | EM85 | 1.25** | PCF86t | $0.75^{\circ}$ |
| EB8CCt 1.00 | ECL86† $0.85^{\circ}$ | EM87 | $1.50^{\circ}$ |  | 1.15* |
| EABC800.55 ${ }^{\circ}$ |  | EN911 | 2.24** | $\begin{aligned} & \text { PCF200t } \\ & \text { PCF201t } \end{aligned}$ | $1.15{ }^{\circ}$ |
| EAC91 ${ }^{\mathbf{0} \cdot \mathbf{5 0}}$ | $\begin{aligned} & \text { EF37At } 2.50^{\circ} \\ & \text { EF39t } 1.60^{*} \end{aligned}$ | EY51t | 0.750 | PCF201t | +1.60* |
| EAF80 1 1.75* | EF40 1.15* | EZ40 | 1.25* | PCF802 | 0.88* |
| EB41 2.00* | EF41 1.20* | EZ41 | 1.25* | PCF805 | 1.44* |
| EB91t 0.40* | EF42 2.00* | E280 $\dagger$ | $0.50{ }^{\circ}$ | PCF806 | 1.44* |
| EBC33 $1.75^{\circ}$ | EF50t 0.60* | EZ819 | $0.50{ }^{\circ}$ | PCF808 | 1.44** |
| EBC41 1.25* | EF80† 0.50 ${ }^{\circ}$ | EZ901 | 0.60* | PCL821 | 0.80* |
| EBC81 $1.10^{\circ}$ | EF83 $1.75^{\circ}$ | GZ32 | 1.25* | PCL83t | 0.92** |
| EBC90 0.75* | EF85 ${ }^{\text {c }}$ 0.50' | GZ33 | $4.00^{\circ}$ | PCL84 $\dagger$ | $0.75{ }^{\circ}$ |
| ERF80 0.50 | EF86t 0.60* | GZ344 $\dagger$ | 1.52* | PCL85t | 0.96* |
| E8F83 E8F89 1.25* $0.45 * *$ |  | K 161 $\mathrm{KT66}$ | 3.50** | ${ }^{\text {PCLL86t }}$ | 0.85** |
| EBL31 ${ }^{\text {2.50* }}$ | EF92t $0.7{ }^{\circ}$ | K166 $\mathrm{KTB8}$ | 5.55* | PCL805 | 85t. |
| ECC40 1.25* | EF98 1-25* | KTW61 | $1.75{ }^{\circ}$ | PD500 | $3 \cdot 60^{*}$ |
| ECC81 ${ }^{\text {¢ }} 0.50^{*}$ | EF1831 $0.70^{\circ}$ | KTW62 | 1.75* | PFL200 | 1.12* |
| ECCB2 ${ }^{\text {¢ }} 0.47^{\circ}$ | EF184 ${ }^{\mathbf{0}} \mathbf{0 . 7 0}$ | KTW63 | 1.75* | PL36t | 1.12* |

## INTEGRATED CIRCUITS

| ARTED |  |  |  |  |  | $\begin{aligned} & 7454 \\ & 7460 \end{aligned}$ | 0.18 0.18 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7400 | 0.16 | 7412 | 0.26 | 7432 | 0.30 | 7470 7472 | 0.35 0.33 |
| 7401 | 0.16 | 7413 | 0.32 | 7433 | 0.36 | 7473 | 0.36 |
| 7402 | 0.16 | 7416 | 0.32 | 7437 | 0.32 | 7474 | 0.40 |
| 7403 | 0.16 | 7417 | 0.32 | 7438 | 0.32 | 7475 | 0.54 |
| 7404 | 0.17 | 7420 | 0.17 | 7440 | 0.18 | 7476 | 0.40 |
| 7405 | 0.16 | 7422 | 0.20 | 7441AN | 0.85 | 7480 | 0.55 |
| 7406 | 0.40 | 7423 | 0.32 | 7442 | 0.72 | 7482 | 0.75 |
| 7407 | 0.40 | 7425 | 0.30 | 7447AN | 1.90 | 7483 | 0.90 |
| 7408 | 0.20 | 7427 | 0.30 | 7450 | 0.18 | 7484 | 1.00 |
| 7409 | 0.20 | 7428 | 0.43 | 7451 | 0.18 | 7486 | 0.35 |
| 7410 | 0.16 | 7430 | 0.17 | 7453 | 0.18 | 7490 | 0.52 |


| 7491 | 0.80 | 74118 | 1.00 | 74144 | 2.50 | 74173 | 1.40 | 74196 | $1 \cdot 20$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7492 | 0.60 | 74119 | 1.50 | 74145 | 0.90 | 74174 | 1.50 | 74197 | $1-10$ |
| 7493 | $0 \cdot 60$ | 74120 | 0.83 | 74147 | 2.00 | 74175 | 0.90 | 74198 | 2.25 |
| 7494 | 0.80 | 74121 | 0.40 | 74148 | 1.75 | 74176 | 1.20 | 74199 | 2.25 |
| 7495 | 0.72 | 74122 | 0.60 | 74150 | 1.60 | 74178 | 1.25 | 76013N1.75* |  |
| 7496 | 0.80 | 74123 | 1.00 | 74151 | 0.85 | 74179 | 1.25 | LM309K 1.50 <br> TAA570 2.30* |  |
| 7497 | 3.00 | 74125 | 0.55 | 74154 | 1.75 | 74180 | 1.15 |  |  |
| 74100 | 1.50 | 74126 | 0.55 | 74155 | 0.85 | 74190 | 1.50 | TAA630 |  |
| 74107 | 0.45 | 74128 | 0.60 0.70 | 74156 | 0.85 | 74191 | 1.50 |  | $50^{\circ}$ |
| 74109 | 0.70 | 74136 | 0.70 0.55 | 74157 | 0.75 | 74192 | $1 \cdot 35$ | TAA700 | $3.91$ |
| 74110 | 0.50 | 74141 | 0.80 | 74159 | $2 \cdot 10$ | 74193 | 1-35 |  | 1.84* |
| 74111 | 0.86 | 74142 | 2.30 | 74170 | 2.30 | 74194 | 1.25 | TBA520 |  |
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\section*{TBA530 1.98 <br> | TBA920 | $2 \cdot$ |
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| No. MVR7918 | HA7918 | TO220 | 75p |
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| $\mu$ A723C ${ }^{\text {¢ }}$ (099 | ${ }_{309 k}$ | $\begin{aligned} & 72314 p \\ & E 1.20 \end{aligned}$ | 38p |

## SWITCHES

No. $161785 \times$ Mains Slide Switches
$\begin{array}{ll}\text { No. S18 } & 4 \times \text { Standard Slide Switches } \\ \text { No S19 } & 4 \times \text { Miniature Pusis to Make }\end{array}$
40p:
No $520 \quad 3 \times$ Mingle nole mounting
No. $521 \quad \begin{gathered}\text { single hole mounting } \\ \text { Push button Switch Pak }\end{gathered}$ $4 \times$ Assorted types multi bank and singles
Latching and no AUDIO LEADS
Order No.
127
2 piugs 5 pin DIN plug to 4 phono
129 Audio lead 5 pin plug to 5 pin DIN plug
$130 \quad 5$ Mirror mage
Mirror mage
metre lead 2 pin DIN plug to 2 pin
DiN inline socket

## AUDIO PLUG AND SOCKET PAKS

Order No.

| Ord |  |  |
| :---: | :---: | :---: |
| S1 | $5 \times 35 \mathrm{~mm}$ Plastic Jack Plugs | 40p* |
| 5 | $5 \times 2.5 \mathrm{~mm}$ Plastic Jack Plu |  |
| S3 | $4 \times$ Std. Plastic Jack Plugs | $50{ }^{\circ}{ }^{\circ}$ |
| S4 | $2 \times$ Stereo Jack Plugs | 30p* |
| S5 | $5 \times 5$ Pin $180^{\circ}$ DIN Plugs | 50p* |
| S6 | $8 \times 2$ Pin Loudspeaker Plugs | 50p* |
| S7 | $8 \times$ Phono Plugs Plastic | 60p ${ }^{\text {c }}$ |
| S8 | $5 \times 3.5 \mathrm{~mm}$ Chassis Sockets ( Switched) | 25p* |
| S9 | $5 \times 2.5 \mathrm{~mm}$ Chassis Sockers (Switched) | 25p* |
| S11 | $2 \times$ Stereo Jack Sockets with instruction leafler for H/Phone connection | 50p* |
| 512 | $5 \times 5$ Pin $180^{\circ}$ DIN Chassis Sockets | 40p |
| S13 | $8 \times 2$ Pin OIN Chassis Sockets | 50p ${ }^{\text {a }}$ |
| S14 | $6 \times$ Single Phono Sockets | 40p" |
|  | P.C. BOARD |  |
| S110 | Mixed Bundle. P.C.B., Fibreglass/paper, single and couble-sided Fantastic value | $75 p$ |

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## SEMICONDUCTOR PAKS

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No 16.31150 Germ. Point contact diodes like OA70/81
No. $16132 \quad 100200 \mathrm{~mA}$ Sil. diodes like OA200
No $16133 \quad 15075 \mathrm{~mA}$ Sil. Fast switching diode like ! N4 148
No. 1613450750 mA Sil. top hat Rect
No $16135 \quad 203 \mathrm{amp}$ Sil. stud Rect.
No 1613650400 mw Zeners D. 0.7 case $\quad$ 40p
No. 1613730 NPN Plastic trans. like BC107/8
No. 1613830 PNP Plastic trans like BC177/8
No. 1613925 NPN trans like $2 N 697 /$
 No. 1614130 NPN trans like 2N706 TO 18 No. 1614330 NPN Plastic trans like 2N3906
No. 1614430 PNP Plastic trans. like 2N3905 No. 16144 30 PNP Plastic trans. like 2N3905
 2N3055

## I.C. \$OCKET PAKS

| No. S66 | $11 \times 8$ pin DHL Sockets |
| :--- | :--- |
| No. 567 | $10 \times 14$ DiL Sockets |

$\mathbf{f 1 . 0 0}$
$\mathbf{f 1 . 0 0}$
$\mathrm{f1.00}$
$\mathrm{f1.00}$
$\mathbf{f 1 . 0 0}$

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| 2N3819 | ${ }_{15 p} \text { FET's }_{2 N 5458}$ | 18p |
| 2 AMP. BRIDGE RECTIFIERS <br> Metal Stud Mounting |  |  |
| No. S45 | 50 V (KBS 005) | 28p |
| No. 546 | 100V (K8S01) | 30p |
| No. 547 | 200V (KBS 02 ) | 34p |
| SILICON BRIDGE RECTS. |  |  |
| S99 | Mixed Pak 2 - 5 Amp. 50-600v. All coded. 4 for | ¢1.00* |
|  | SIMILARIN4000 SERIES |  |
| No. $\mathrm{S4} 1$ No. $\mathrm{S42}$ |  | ${ }_{60 p}^{60 p}$ |
| No S43 | 18 Like $\operatorname{Na} 4003$ (1A 200 V ) | 60 p |
| No. S44 | 15 Like IN4004 (1A/400V) | 60p |

## 

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| CONPONENT PAKS |  |
| :---: | :---: |
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| 16170 | 50 metres Single strand wire assorted wire |
| 16171 | 10 Reed switches |
| 16172 | 3 Micro switches |
| 16176 | 20 Assorted electrolytics Trans types |
| 16177 | 1 pack Assorted Hardware |
| 16179 | 20 Assorted tag strips and panels |
| 16180 | 15 Assorted control knobs |
| 16184 | 15 Assorted Fuses 100 mA 5 amp |
| 16188 | $60 \frac{1}{2} \mathrm{~W}$ resistors mixed values |
| 16187 | 30 metres stranded wire assorted colours |
| S100 | $120 \frac{1}{4}$ watt resistors Pre-formed 1978 Prod Our mix |
| S101 | 120 ; watt resistors. Pre-forme |
| S102 | 1978 Prod. Mixed values |
|  | Range 100 othms - 10 meg |
| S103 | $220 \frac{1}{2}$ watt resistors |
|  | Aange 100 ohms - 10 meg |
| S104 | 60 Low ohms $\frac{1}{3}$ watt resistors 10-100 ohms |
| \$105 | 40 Low ohms $\frac{1}{2}$ watt resisto |
|  | 10-100 ohins |
| S106 | 25 Mlixed wirewound resisto |
| S107 | 20 Tantatumbead caps 0.22-100mF Our mix |
| 5108 | High quality electrolytics 10 m |
|  | 500 mF voltaze range 1550 Our mix: 40 for |
| 16204 | C280 Pak Contains 50 metal |
|  | caps |

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Slider 40mm TRAVEL
OrderNo.


S90 Wirewound Pots Linear 1 Watt rating Mixed useful values 5 for $\mathbf{£ 1 . 0 0}$ CARBON TYPES
591 Car Radio type Dual \$wirched Pot 100 KLinting switched
2.5 K in

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| :---: | :---: | :---: |
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| No S56 | 20 mixed vaiues 400 mW Zener diodes $11-33 \mathrm{~V}$ | £1.00 |
| No. S57 | 10 mixed values $1 W$ Zener diodes 3-10V | f1.00 |
| No. S58 | 10 mixed values IW Zener diodes 11-33V | f1.00 |
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| S97 | BD3712 Amp 1-2w. 60vceo Hte 40 400. Case TO92 with heal tab 5 for | 60p" |
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AC187K
AC188
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## 162 MP AF 139 AF 239

AF239
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$\mathrm{BC}_{10}$
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$\begin{array}{ll}\text { 16p } & \text { BC184 } \\ \text { 26p } & \text { BC184L }\end{array}$$\begin{array}{ll}\text { 20p } & \text { BC212L } \\ \text { 80p } & \text { BC213 }\end{array}$$\begin{array}{ll}9_{p} & \text { BF19 } \\ 9_{p} & B F 197 \\ 9_{p} & B F 200 \\ 9_{p} & B F \times 29\end{array}$
$\begin{array}{ll}\text { 80p } & \text { BC213L } \\ \text { 30p } & \text { BC214 } \\ \text { 30p } & \end{array}$
$\begin{array}{llll} & & \text { 12p } & \text { TIP42C } \\ \text { TIP2955 } \\ \text { 10p } & & & \text { TIP3055 } \\ \text { 10p } & \text { MPSAO5 } & \text {-22p } & \text { ZTX107 } \\ \text { 10p } & \text { MPSAO6 } & \text {-22p } & \text { ZTX108 }\end{array}$
BC147
BC148
BC149

| To. |
| :--- | :--- |

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TO those readers who have suffered delay and disappointment in obtaining their copies of Practical Wireless of recent months, we offer our sincere apologies. In some of the more remote parts of the UK, late publication and some hiccups in distribution have combined to produce a particularly trying situation. We hope that these difficulties are now being overcome and that the magazine will be on sale by the specified date each month.

Another disappointment for some of our readers, is being unable to obtain a copy of an issue because every one has been snapped up, and their newsagent just cannot get any more. Worse still is picking up a casual copy of $P W$ only to find that an interesting series has been running for some months and they can't get hold of the previous issues. We always do what we can to help someone in this predicament, and have recently increased the number of magazines put aside each month for our back numbers service (see page 1). Unfortunately though, many issues are now completely out of print, and only very limited stocks exist of some others. The table below summarises the situation as we went to press.

|  | 1979 | 1978 | 1977 | 1976 | 1975 | 1974 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| JANUARY |  |  |  |  |  |  |
| FEBRUARY | $\ddots$ |  |  |  |  |  |
| MARCH |  |  |  |  |  |  |
| APRIL |  |  |  |  |  |  |
| MAY |  |  |  |  |  |  |
| JUNE |  |  |  |  |  |  |
| JULY |  |  |  |  |  |  |
| AUGUST |  |  |  |  |  |  |
| SEPTEMBER |  |  |  |  |  |  |
| OCTOBER |  |  |  |  |  |  |
| NOVEMBER |  |  |  |  |  |  |
| DECEMBER |  |  |  |  |  |  |

If you want to make sure of getting your copy of $P W$ each month, either place a regular order with your newsagent, or take out a subscription--again, see page 1 for details.


## Debbie Chapman, Copy Typist

A secretarial course which she completed immediately after her normal schooling at the Purbeck School, Wareham, led Debbie into her first job which, luckily for us, was with PW.

She was born in Kenya, where her father was serving in the Army, and most of her life so far has been spent on a "world trip" with her father and the rest of the family. Dad is an instructor with the 3rd Royal Horse Artiliery at Bovington Camp, specialising in tanks.

Mum also works for the Army, and Debbie is engaged to a Police Constable-she certainly maintains an interest in uniforms!

Among her other interests, which include swimming and looking after children, she quotes cooking as the most important. Future husband Martin seems to have it made in no uncertain terms-good food and a willing baby sitter while he's out working can't be bad.

## Well done Erzsebet

The 1978 Girl Technician Engineer of the year is Mrs Erzsebet Kibble (see News, October 1978): the award was announced by The Secretary of State for Education and Science, The Rt Hon Mrs Shirley Williams, MP, at a meeting of The Institution of Electrical and Electronics Technician Engineers (IEETE) in London on Monday, 4 December.


Erzsebet, aged 25, is an Assistant Test Manager with Thorn-Ericsson Telecommunications (Sales) Ltd. Born in Hungary, Erzsebet is married to an Englishman and lives in Woking, Surrey. She joined her present employers in 1974 and gained the Higher National Certificate in Electrical and Electronic Engineering in 1976.

The Award, accompanied by a cheque for $£ 250$ and an inscribed rose bowl, is sponsored by The Caroline Haslett Memorial Trust and the IEETE. Its aim is to focus attention on electrical and electronic engineering as a worthwhile career for women. IEETE, 2 Savoy Hill, London WC2R OBS. Tel: 01-836 3357

## New catalogues

Barrie Electronics Ltd. inform me that their latest catalogue is available. Although the 32 page catalogue lists a good deal of Barrie's stock, they have many other components, too numerous to list. They also invite readers to contact them if they need assistance with purchasing any difficult items.

The catalogue is obtainable for $15 p$, from: Barrie Electronics Ltd., 3 The Minories, London EC3N 1BJ. Tel: 01488 3316/7/8.

Suhner Electronics, the UK subsidiary of the Swiss company Huber \&

Suhner, have recently issued 2 new catalogues covering just part of the range of connectors they manufacture.

1. High Voltage Connectors; this 16 page catalogue covers 3 different series of coaxial high voltage connectors with 50 ohm nominal impedance and bayonet coupling mechanisms. In all a total of 80 different connectors.
2. RF Connectors Series TNC; this comprehensive 28 page catalogue details Suhner's range of TNC mediumsized coaxial cable connectors.

Copies of these catalogues are available, free of charge, from: Suhner Electronics Ltd., Telford Road, Bicester, Oxfordshire.

Three new parts of the Mullard Technical Handbook are now available. They are: RF power devices-Book 1, Part 2-£2.00; Thyristors and triacsBook 1, Part 5-£1.50; Loudspeakers, television assemblies and modulesBook 3, Part 5-£2.00.

The Mullard Technical Handbook is broken down into twenty-one different parts which can be purchased separately-the prices depend on the contents and include P\&P.

The handbooks and a leaflet giving further details is available from: Central Enquiry Handling Unit, Mullard Ltd., New Road, Mitcham, Surrey CR4 4XY.

## Operation Drake

Radio communications equipment, supplied by companies in the GECMarconi Electronics group, will play a big part in Operation Drake-the two year round-the-world voyage by parties of experienced explorers, scientists and 24 young explorers in the 150 ton brigantine Eye of the Wind, which left Plymouth on Wednesday, 8 November, to celebrate the 400th anniversary of Sir Francis Drake's circumnavigation of the globe.

Marconi Marine is providing one of its new 400 watt Transocean/Pacific s.s.b. radiotelephones to satisfy the vessel's requirements for m.f. and h.f. communications with shore-stations round the world. This equipment is powered by the vessel's a.c. mains supply.

In case of generator failure on board, Marconi Marine is also supplying a 24 V d.c. operated Falcon II, a m.f./h.f. s.s.b. radiotelephone as a back-up set for the main equipment.

When working within 40 miles of the coast, the primary ship-to-shore
communications will be supplied by an Argonaut S v.h.f. radiotelephone.

For emergency purposes Marconi Marine is also to provide a Survivor II survival craft radio equipment.

Communications in the v.h.f./f.m. bands will be provided by three UK/VRC353 transceivers for the overland expeditions.

## PW Gillingham

It has come to our notice that readers are having difficulty obtaining the 1.280 MHz crystal for the PW Gillingham Short-wave Receiver Frequency Readout, published in the October 1978 issue.

A special arrangement has been made to supply these devices, at an inclusive price of $£ 4.15$, from P. R. Gollege Electronics, Merriott, Somerset TA16 5NS. Tel: (0460) 73718.

## Project Index

A new index of electronic projects has been compiled by M. L. Scaife, Principal Librarian (Technical) of North Tyneside Libraries and Art Department.

The index provides a descriptive guide to over 2500 projects published between 1972 and 1977, in journals such as Practical Wireless, Practical Electronics, Wireless World and Television Magazine.

The project is not intended as a profit-making venture, as the following inclusive prices indicate: 1-2 copies£1.50, 3-6 copies-£ $1.40,7-10$ copies- $£ 1.35$, over 10 copiesspecial rates.

Copies of the index are obtainable from: M. L. Scaife, Central Library, Northumberland Square, North Shields, Tyne and Wear NE3O $10 U$. Tel: (08945) 82811.



## OSO? GB2RN

The amateur radio station aboard HMS Belfast moored in the pool of London, between Tower Bridge and London Bridge, has been granted the use of the special callsign GB2RN when the ship is open to the public. Summer hours 1100 to 1800 , winter hours 1100 to 1630, all British local time.

The station is interested in establishing schedules with other museum and special interest stations worldwide, these and other stations who would like to arrange skeds, please contact Don Walmsley G3HZL, 153 Worple Road, Isleworth, Middlesex TW7 7HT, England.

All h.f. bands from 1.8 to 28 MHz are covered on c.w. or s.s.b.; it is hoped to have $R T Y$ in the near future.

G4HMS will be operational outside of the stated hours.

## LCDs

Two liquid crystal displays (I.c.d.s) will be available from Mullard in quantity early in the year. These devices join the Company's already established ranges of solid-state and plasma display devices. This announcement is the result of a long-term development programme which not only involved investigation into different techniques but also into basic materials in order to ensure that the displays would give continuing high-standard performance coupled with assured long-term reliability.
The technology selected for the Company's I.c.d.s was field-effect twisted-nematic. This technology has been in the market place for some time, but the Mullard I.c.d.s will incorporate the latest developments in this field. In particular, nematic liquid crystal materials have been developed to provide high standards of chemical stability and special precautions have been taken to eliminate chemical reaction between the different materials used. Also, to ensure satisfactory viewing, the l.c.d.s incorporate the results of a great deal of research into viewing angles, colour balance and the brightness ratio of the character to the background areas.

Many different types of l.c.d.s are in use today in watches, clocks, calculators and a great deal of instrumentation where their low power consumption, compared to that of l.e.d.s, enable them to be driven
directly by m.o.s. devices.
The Mullard l.c.d.s were designed specifically for time displays and provide a 4 -digit read-out with a colon between the second and third digits. Each digit is 12 mm high and is formed from 7 segments. The first of these l.c.d.s, type LTC001R, is designed for use in the reflective mode: that is, it collects ambient light and reflects it back to the viewer. The second I.c.d., type LTC001T, is used in the transmission mode with a light source behind the display.
Mullard Limited, Mullard House, Torrington Place, London WC1E 7HD.

## GB3WM

The Home Office have granted the special callsign GB3WM to the Wireless Museum at Arreton Manor, near Newport, Isle of Wight. Arreton Manor is the home of The Count and Countess Slade de Pomeroy, the Manor is open daily from Easter until November, and also by appointment during the winter months.

The Wireless Museum contains a unique collection of very early radio and television receivers, including a 30-line Baird Televisor with spinningdisc mechanical scanning.

It is now planned to build an exhibition short-wave experimental transmitting station, as used by radio amateurs before the last war, and display it side-by-side with a modern piece of equipment, showing a direct comparison between the old type of rack-mounted equipment and today's 'state of the art".
All amateur bands will be used, with both c.w. and s.s.b. on the high frequencies and f.m. on two-metres (via GB3SN, the Alton, Hampshire repeater).
Official station operators will be Messrs. D. Byrne G3KPO (Museum Curator), A. R. Williams G3KSU, D. E. Denny G3ZQE, A. Wakeley G3MAD, R. W. Fisher G2DZN, K. B. Pearse G3MLC, F. D. Cawley G2GM, H. Childs G3IOW, L. Critchley G3EEL, D. Hoult G400 and W. Carter G2NJ.

A special OSL card will be printed and issued to all initial QSOs. Shortwave listeners are also invited to send in report cards, for which OSLs will be exchanged-all via the RSGB Bureau, or on receipt of an s.a.e.
D. Byrne G3KPO, The Wireless Museum, Arreton Manor, near Newport, Isle of Wight.

## TMMETRON

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$\mathbf{£ 7 4 . 9 5}$


## CASIO CHRONOGRAPH

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## Radio 3

Sir: While admiring the ingenuity of the BBC Engineering Department in replacing Radios 1, 2, and 4 transmitters, I must complain about the allocation of 1215 kHz for Radio 3 . This frequency is only fit for its original purpose-a back-up for I.w. transmissions in big cities. The map on page 41 of 18-24 Nov 1978 issue of Radio Times confirms this. The 1215 kHz frequency is subject to hideous phase distortion here, and swamped by Radio Albania after 3 pm .

I am primarily concerned with conditions during daylight hours. Most people have v.h.f. at home, although that is difficult in some parts of this town, and v.h.f. car radio is ineffective here.

I am waiting for the howls of protest, when we have the first "Test Match Special" on m.w. only. I have had a long correspondence with the BBC and I am now hopeful that they will, after 23 November, try to improve Radio 3 outlets. Small relays could go on an international common frequency-or even a borrowed one, as at Cambridge (that is not on the Geneva list). What we urgently need is one more frequency to be used by alternative main stations as for Radios 1 and 2. We still have, as yet unused, an allocated frequency of 227 kHz , although it may be liable to interference.

It is a scandal that the Home Office did not even ask for extra channels at Geneva (see Richard Last, page 12, Daily Telegraph, 13 November 1978). How did small countries like Belgium, Holland and Eire, manage to get new valuable long and medium wave allocations, all more efficient than the 648 kHz , which is now lost to us, for the BBC European Service.

I hope that all your readers who enjoy Radio 3, especially in cars, as I have done for years, will make their views known to the BBC as soon as possible. Then, perhaps, some improvements will occur.

## Dr H. S. Brodribb <br> St Leonards-on-Sea

## Any offers

Sir: I have for sale approximately 150 copies of various electronic magazines, such as PW, PE, EE and WW. Some of the issues date back to 1946/7, if anyone is interested, I would accept any reasonable offer plus postage.

I also have a quantity of electronic bits and pieces (valves, transistors, transformers, motors etc.), for which all I require is the postage.

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In recent years, Sound-to-Light (STL) units, in various forms, have become extremely popular for discos, pop concerts, parties, shop displays and so on. In its simplest form, a STL unit consists of three of more CR filters fed from the loudspeaker terminals of an audio amplifier, feeding transistors which trigger triacs (or thyristors) via pulse transformers. Each channel has its own level control, and there may be a "master" control as well. This form of STL has several drawbacks; the sensitivity is low (an audio signal of $2-3 \mathrm{~V}$ may be required) and continual adjustment is required to compensate for changes in volume and tone. The apparent separation between channels is poor, due to the simple filters used: the unit generates interference spikes on the mains, since the triacs may be triggered at any point on the mains cycle, and it uses several heavy and/or expensive transformers.

The STL unit to be described overcomes all of these disadvantages; it uses interference-free "zero-voltage switching", so that the triacs are triggered only when the
voltage across them is very low; it has active filters for excellent separation between channels and automatic gain control (a.g.c.) amplifiers to eliminate manual adjustments; the sensitivity is approximately 700 mV to control 1 kW of lights per channel, and it uses only one inexpensive transformer. Due to the virtual elimination of transformers, this design is little more expensive to construct than the conventional version described above.

## The Circuit

As can be seen from the simplified block diagram (Fig. 1), the circuit comprises several blocks each with a welldefined action. Each block is identified in the circuit diagram (Fig. 2). The signal from the loudspeaker circuit of an audio amplifier is fed via transformer T1. This provides isolation between the amplifier and the remainder of the circuit, which is at mains line voltage. Consequently, extreme caution is required when constructing the circuit.



Fig. 2: Complete circuit diagram (excluding options)

Note that T1 is actually a miniature mains transformer, which has good isolation and adequate frequency response for this application. The audio signal then enters an a.g.c. amplifier which allows the STL to operate with inputs from 700 mV r.m.s. $(0 \cdot 12 \mathrm{~W} / 4 \Omega$ to $25 \mathrm{~V}(150 \mathrm{~W} / 4 \Omega)$. This consists of a voltage controlled attenuator $\mathrm{R} 16 / \mathrm{Tr} 3$, followed by a X30 amplifier IC2a, with a control voltage derived from the rectifier circuit D6/D7.
If more sensitivity is required, R13/R14 may be reduced to, say, $1 \mathrm{k} \Omega 1 \mathrm{~W}$, but this loads the audio amplifier somewhat more, and increases the possibility of noise affecting the lights. If a high voltage source is used (e.g., 100 V lines) R13/14 should be increased to $15 \mathrm{k} \Omega$ 1W. Note that a stereo input is provided; if mono only is
required, both inputs should be tied together, alternatively a mono jack plug in the (stereo) jack will work with slightly reduced sensitivity.

After processing in the a.g.c. amplifier, the audio signal is fed to three active filters: low-pass (bass, IC2b), bandpass (midrange, IC2c) and high-pass (treble, IC2d). These have their cut-off frequencies chosen to give subjectively good separation between channels with most types of music from a variety of sources from a portable transistor radio to powerful discotheque systems. The output of each filter is connected to one of three identical detector/a.g.c. circuits constructed around a quad comparator integrated circuit.

For simplicity, consider the circuit connected to the low-pass filter (IC1b). If there is no audio signal, the comparator output will be low due to the bias voltage from D10 and R32, and no trigger pulses will be applied to the
triac. Now, if a steady audio tone is applied, capacitor C14 will rapidly charge via D9 and the comparator output will go high. This allows trigger pulses, timed to coincide with the points in the mains cycle where the voltage is zero, to be applied to the triac. Thus, the triac will turn on when the voltage across it is small and the appropriate lights will come on without any interference being generated. However, C15 slowly charges via D8 and eventually the voltage across it exceeds that on C14. The comparator switches again, turning the lights off. With more complicated dynamics in the audio signal the situation is more complex, but it can be seen that this circuit provides



Fig. 3: (a) (top) Addition of pushbutton and switched override facilities. (b) Half/full power switching

A rear view of the prototype, showing input jack, fuses and eight-way output socket
an a.g.c. type action. The preset VR1 allows adjustment of the operating point of the circuit for optimum operation.

The detector circuits described above are fed with "zero voltage" pulses from the pulse generator circuit Trl/ICla. The mains input is clipped by D3, and an inverted squarewave is produced at Trl collector. The positivegoing edges of both are differentiated by CR networks, and trigger a comparator IC a providing short positivegoing pulses at Tr 2 collector. The positive supply for the whole unit is provided by D2 and C1.

## Circuit Modifications

The circuit described so far is capable of forming a simple but effective "no frills" STL unit, with no external controls whatsoever! If, like the writer, you prefer more facilities, several additions can be made, some of which are provided for on the printed circuit board described later.

First, a mains on-off switch can be provided-useful for quick changes between lighting effects. Switches can be provided (Fig. 3(a)) to override the triacs on each channel, switching the lights permanently on. If the switches have built-in neon indicators, these can be connected to provide monitoring of the light display. These switches must be capable of carrying the full load current. However, low power switches or push-buttons may be used with a simple diode gate (also shown in Fig. 3(a)), providing fast, noisefree switching. This method is particularly suitable for push-buttons, allowing the lights to be rapidly flashed under manual control. These additions are included in the version described in the constructional details and positions for the extra components required are included on the p.c.b.

Other improvements include the addition of a half/full power switch (Fig. 3(b)). This short-circuits Tr1, so that only half of the differentiated pulses reach the comparator. Thus, the triacs are only triggered on positive half-cycles, so the lights will be less bright. In a similar manner, an audio on/off switch can be added, allowing the override push-buttons to be used alone. This switch short-circuits R15, preventing audio signals from reaching the a.g.c. amplifier.

More ambitious modifications include making presets VR1-VR3 front panel controls (for the "knobs with everything" man!) or even providing controls with a switched "automatic" position. If this is tried, the controls must be well insulated. Also, a sequencer facility can be added, using transistors in place of the push-buttons in Fig. 3(a), or a dimmer control can be added using a 555 oscillator. These modifications are left to the more experienced constructor.

## Construction

WARNING: Most of the circuit board and wiring is at MAINS LIVE VOLTAGE and EXTREME CAUTION is necessary when testing or adjusting the circuit. The use of an Epoxy Glass Fibre p.c.b. is strongly recommended; Veroboard, etc., is NOT suitable, as it has insufficient intertrack voltage rating.


Fig. 4 (above): Printed circuit board track pattern, shown full size.

Fig. 5 (below): Component layout



Internal view of the prototype unit

The p.c.b. track pattern is shown in Fig. 4 and the component layout in Fig. 5. Constructors should have little trouble in assembling the board provided the specified components are used. The usual precautions against the effects of static electricity should be taken when handling the two c.m.o.s. integrated circuits. It is safest to use sockets as specified in the components list, and insert the i.c.s only when the remainder of the components have been fitted to the p.c.b. Note that the wirewound resistor R1 gets hot in operation and should be mounted away from the board. Also, the triacs may require larger heatsinks if the unit is constructed in a small, enclosed box. If a common heatsink is used, mica washers and insulating bushes will be required. If the push-button override facility is not required (this is shown on the layout diagram), $D_{A}$, $D_{B}$ and $R_{A}$ (three of each) may be omitted and $D_{A}$ replaced by a wire link. Finally, the p.c.b. has extra holes to allow the presets to be replaced by fixed resistors when the optimum values have been found.

The wiring of the unit is not critical and any convenient layout may be used. For guidance, the wiring used in the writer's unit is shown in Fig. 6, drawn flat for clarity. All mains wiring should, of course, use thick wire, say $40 / 0.2 \mathrm{~mm}$, remembering that some leads can carry up to 12A! Note that diode D1 is mounted off the board and resistors R13-R15 are mounted on the input jack JK 1. Any convenient case may be used, but the circuit must be securely protected from prying fingers. All metal parts except the heatsinks must be efficiently earthed. The writer's unit was constructed on a metal chassis in a wooden sleeve as shown in the photographs, but alternative construction methods could easily be used.

Fig. 6: Wiring diagram of a unit incorporating options of Fig. 3(a)



## Testing and Setting Up

After construction is completed, all wiring should be double-checked and all presets set to their central position. Connect a set of lights and switch on. The lights may flicker briefly then they should remain off. The push button switches (if fitted) should be tested at this point; this checks the operation of the triacs. If all seems well, switch off and connect an audio source. Setting up is best done with a moderately large input signal (2-3V r.m.s.). Switch on again and the lights should show some response. The presets should be adjusted, with an insulated tool, to give the best results with various types of music; this need only take 15 minutes or so. Finally, vary the volume over a wide range and check that the display is similar at different volume levels. The a.g.c. amplifier takes a second or so to "catch up" after rapid changes in volume, so allow a little time between adjustments, etc

Any faults found are likely to be due to incorrect assembly (e.g., short-circuits on the board, reversed tantalum capacitors), faulty semiconductors or "dry" joints. Note that it is no use trying to test the unit using only the neon indicators in the override switches (where fitted), instead of external lamp loads. The neons do not draw sufficient current to hold the triacs on after they have been triggered.

In cases where there is an exceptional amount of electrical noise on the mains supply from other sources, the lights may flicker even when no audio signal is present. If this is a problem, a $0.1 \mu \mathrm{~F} 1000 \mathrm{~V}$ capacitor across the mains supply should cure it (see Fig. 6); alternatively, or in extreme cases additionally, the sensitivity may be reduced slightly by increasing the value of R17.

The unit described has been very reliable in use, and several prototypes have been constructed. With care, constructors will have a Sound-to-Light unit of professional standard, capable of long service.


## STEP TONE GENERATOR



This circuit produces a sound which relates to defined increments stepping up and down a musical scale. No shift register or counter is required to produce each step, and with the exception of the peripheral components the circuit consists of only two 555 timers, an f.e.t., and a unijunction transistor.

One 555 is used as an astable driving a loudspeaker, with the other functioning as a triangular waveform generator providing a control ramp for feeding back into the astable.

Increments are provided by spikes from the unijunction at intervals of $100 \mu \mathrm{~s}$, which are impressed onto the triangular waveform. The output of the astable is fre-
quency modulated at the same rate, and as the triangular waveform changes, a series of loud beats are produced, running up and down the musical scale. The f.e.t. acts as a buffer to prevent loading of C 1 . This circuit could be modified to become a "blind man's d.v.m." or a relatively sophisticated voltage to frequency noise scrambler.

> D. Brown GM8FFH,
> South Hawthornhill,
> Dumbarton.

## CAR CASSETTE POWER SUPPLY

This simple circuit uses a 7805, which is a compact device used mainly with TTL i.c.s and fed with an ample input of at least 8 V will produce a stabilised output of 5 V at up to 1 amp , provided it is mounted on a heat sink. The minimum of smoothing is required, and the overall circuit will provide 5.6 V to power a cassette player from a car battery.

Only a handful of components are needed and R1 was selected such that no voltage drop occurred at maximum volume and of course, the voltage never rises beyond 5.6 V without a load. There is, therefore, little chance of the device becoming faulty and damaging the cassette recorder. If a higher output voltage is needed, another diode


can be connected in series with D1 giving 6.2 V but for safety, the existing output is sufficient for most applications.

The output can either be connected directly to the battery connectors in the recorder or alternatively a low voltage power plug can be used to connect externally to the recorder if a socket is fitted.

The unit will work with cars using either positive or negative earth but if a metal box is used as a case, care must be taken to prevent it coming into contact with any part of the car body, especially if a positive earth is used.
M. Burrell, Halstead, Essex.

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The development of the Compact Cassette system from its inception to the present day is reviewed by Gordon J. King

After completing the assembly of the receiver as shown in Fig. 5, a careful visual check should be carried out on all wiring, the unit connected to a 12 V d.c. supply and the supply current measured. This should be approximately 75 mA in the absence of a signal. If a signal generator is available, this should be connected to the aerial socket via a suitable attenuator, and set to $455 \mathrm{kHz}( \pm 1 \mathrm{kHz})$ with internal modulation. A tone should be heard when the receiver is switched to give a.m. reception, and it should be possible to peak IFT1 and the filter IFT2 for maximum output. If necessary, increase the attenuation between the signal generator and the receiver progressively as the i.f. circuits are peaked. This alignment is best done with a relatively weak signal since the a.g.c. action will, to some extent, mask changes in the output level for strong input signals.

An input to the receiver of about 1 mV at 455 kHz will be all that is required in order to produce a strong audio output signal if the i.f. stages are working correctly. If a signal generator is not available, the i.f. stages may be aligned by tuning the receiver for a weak signal and then adjusting IFT1 and the filter for maximum output. In either case, the alignment process is greatly simplified by reference to the signal strength meter. To adjust the meter, VR4 is first set to mid-position and, with the aerial disconnected, VR5 is then adjusted to zero the meter. VR4 may later be set to calibrate the meter. Constructors should note, however, that there may be a slight interaction between these two controls and the setting-up procedure may have to be repeated whenever the calibration is being altered.

After completing the i.f. alignment, the r.f. tuned circuits of L1 and L2 should be aligned. The signal generator should be set to 1.45 MHz with internal modulation and VC1 set for maximum capacitance (i.e., vanes fully meshed). VC2 should be set to mid-position and the core of L2 adjusted until a strong signal is heard. The core of L1 is then peaked for maximum signal. The signal generator is then set to 3.5 MHz and VC1 adjusted to near minimum capacitance (i.e., vanes almost fully open) until a strong signal is again heard. The core of L1 should again be peaked for maximum signal but the core of L2 should
be left alone. If, however, the 3.5 MHz signal is not located, the core of L2 may be adjusted until a signal is heard at minimum capacitance. This establishes the high frequency end of the receiver's tuning range. There will be a noticeable difference in the settings of the core of L1 at 1.45 MHz and 3.5 MHz , due to the "tracking" error. With VC2 still set to mid-position, the signal generator is tuned to 2.5 MHz and a strong signal located at approximately mid-position of VC1 (i.e., middle of the tuning range). The core of L1 is peaked for maximum, when its position should be roughly mid-way between the 1.45 MHz and 3.5 MHz positions.

If a signal generator is not available, signals on known frequencies can be used as an aid to alignment. These could include broadcast signals at around 1.5 MHz , coastal radio stations on frequencies around 2 MHz , and the 2.5 MHz standard frequency and time signal. The vernier dial can be calibrated by first removing the scale, painting it with a matt-finish white or silver paint, and then using dry transfers to provide markings every 100 kHz from 1.5 MHz to 3.5 MHz . Alternatively, a somewhat simpler solution is to draw a calibration curve for the receiver and to refer to this whenever exact frequency readout is required. A typical calibration curve is shown in Fig. 6.

## BFO

In order to align the b.f.o. transformer, IFT4, a strong carrier, preferably with no modulation, should be selected. With the receiver switched for a.m. reception, VC1 should be carefully tuned to provide maximum indication on the signal strength meter. The mode switch, S1, is then set to the s.s.b./c.w. position and, with VR3 set to mid-position, the core of IFT4 is adjusted for zero-beat.

A beat note may not, in some cases, immediately be heard since the heterodyne produced may lie outside the audible range. If this is the case, adjustment of the core of IFT4 should readily produce an audible note. The quadrature detector tuned circuit, IFT3, is best adjusted using a known f.m. signal of preferably 3 kHz peak devia-
backed off for best results. The facility to receive f.m. (which is very rarely used at h.f.) was thought to be a useful addition to a receiver which can potentially be used as a tuneable i.f. in conjunction with a v.h.f. converter. The f.m. facility provides for the reception of narrow-band f.m. signals of typically not more than 3 kHz peak deviation. Limiting action is automatically provided for strong signals. The f.m. detector will, however, not perform well with wide-band f.m. signals and, should this be an important consideration, constructors are advised to replace the i.f. filter IFT2 with a comparable type having a wider bandwidth (e.g., the CFT455B which has a nominal 8 kHz bandwidth and is pin-compatible).

The receiver is eminently suitable for either fixed or portable use. In each case the receiver will benefit greatly from the use of a properly designed aerial system and tuning unit or preselector. This ensures optimum performance and also helps to reject signals on the image channel. These sometimes appear as breakthrough of strong h.f. signals spaced by twice the intermediate frequency (i.e., 910 kHz ) above the wanted signal. A particular problem associated with sensitive low- and medium-frequency receivers is that of line-timebase interference from nearby television receivers. There is often little that can be done to prevent this nuisance, but in exceptional circumstances it may be necessary to site the aerial some distance away from dwellings and to employ a buried coaxial feeder with its screen earthed at each end.

For mobile use, a simple base-loaded whip aerial should be adequate although, here again, some form of preselector or tuning unit will give vastly improved results.


Fig. 6: Tuning calibration curve, used as an aid to alignment


Fig. 7: The basic circuit of the image channel rejector

Fig. 8: An input tuned circuit using a ferrite rod


## Aerial Matching

The additional tuned circuit shown in Fig. 7 will provide an extra 20 dB of image channel rejection. The variable capacitor VC should have a maximum value of about 500 pF . Ll consists of 100 turns of 36 s.w.g. enamelled copper wire wound in two layers on a 4.8 mm diameter former fitted with a dust core, base and screening can. L2 consists of 10 turns of $30 \mathrm{~s} . \mathrm{w} . g$. over-wound on L1. C $\mathrm{C}_{\mathrm{x}}$ can be 47 pF for short aerials or 15 pF for a long-wire aerial.

If desired, this form of circuit can be modified to use a ferrite-rod aerial, as shown in Fig. 8, though it should be remembered that such an aerial is directional. Here, VC should have a maximum value of 208 pF . L1 is 22 turns of 30 s.w.g. enamelled copper wire wound at one end of a 180 mm ferrite rod (see Fig. 9), and L2 is 5 turns of 30 s.w.g. enamelled copper wire over-wound on L1. The ferrite rod should be mounted on pillars about 50 mm long, either above or to the rear of the receiver.

An aerial matching unit provides an excellent method of improving the r.f. selectivity of the receiver, thus adding considerably to the image channel rejection, and also acts as an impedance match. In the circuit of Fig. 10, Tr1 is operated as an emitter follower which exhibits a voltage gain of slightly less than unity with a high input impedance and a low output impedance. The circuit can be constructed on a small piece of Veroboard or matrix board. Component details for the input tuned circuit are shown in the table.

## Tuned Circuit Details for the Aerial Matching Unit



## $\star$ components



Fig. 9: Details of the ferrite rod aerial

Fig. 10 (below) the circuit diagram of the aerial matching unit


## Power Supply

The receiver will give many hours of operation from its internal battery supply which consists of eight U11 cells or equivalent. However, a mains power supply is a very useful addition to any receiver which is likely to be used for any length of time at a fixed location. The circuit diagram of a suitable supply is given in Fig. 11, with p.c.b.


Fig. 11: The circuit diagram of the mains power supply unit

track pattern and component layout shown in Figs. 12 and 13 respectively.

A centre-tapped mains transformer T1, and a full-wave rectifier (diodes D1 and D2) are used to build up a d.c. voltage across the reservoir capacitor C 1 . The smoothed output is stabilised by the series regulator transistor $\operatorname{Trl}$, driven by a d.c. amplifier based on Tr 2 . Control of the output voltage from the unit is achieved by supplying the base of Tr 2 from an adjustable tap on the potential divider


Copper track pattern and layout of the power supply unit shown full size
formed by R4, VR1 and R3. Zener diode D3 provides an emitter-reference potential for Tr2. Changeover from a.c. mains to battery power is automatic in the event of a supply failure. This is achieved by the steering diodes D4 and D5. When the mains supply is present, the potential at point " $A$ " will exceed that at point "B" (see Fig. 11). This will reverse-bias D4 and forward-bias D5, effectively open-circuiting the battery supply. When the mains supply is removed, either by accident or design, D4 will be forward-biased and D5 reverse-biased, supplying battery power to the receiver.

The power supply is straightforward to build and only one adjustment is necessary. Before connecting the output of the supply to the receiver board, the voltage at point "A" should be set to +13 V by adjustment of VR1. In case of any problems, typical test voltages as measured using a $20 \mathrm{k} \Omega / \mathrm{V}$ multimeter on the 25 V range are indicated on the circuit diagram.

The a.c. mains plug should be removed from the supply when the receiver is not required for operation over an extended period, since the mains unit remains working independently of the front panel on/off switch.

# THE <br> <br> IU 'WINTON' <br> <br> IU 'WINTON' Stereo Amplifier 

 Stereo Amplifier}

The Winton amplifier has been designed primarily for the home constructor who would like a high fidelity amplifier equal to the best commercial design available, but at a price that enables it to be built within a sensible domestic budget, the total outlay being in the region of $£ 110$ for the complete project.

All the components are available and no special test equipment is needed for testing or setting up. The amplifier uses the very latest techniques and is capable of a standard of performance at least equal to the best commercial designs available with similar power ratings. In some aspects of its design it will outperform other amplifiers costing very much more.
Power f.e.t.s are used in the output stage and allow a wider power-bandwidth response with lower distortion than could be obtained if conventional bipolar output transistors were used. In the control unit section, bi-f.e.t. op.amps are used. These have an improved slew rate over the more common types, as well as much lower distortion. In the disc (magnetic) pre-amplifier, a three stage circuit using ultra low noise transistors enables noise figures of -68 dB unweighted relative to 50 W output to be obtained, with distortion so low that the input (normally 3 mV ) has to be increased to $140 \mathrm{mV}(+33 \mathrm{~dB})$ before the distortion even reaches $0 \cdot 1 \%$. A full circuit description will be given later.

On the front panel the Winton amplifier has controls for Volume, Balance, Bass, Treble and push buttons for selecting, Disc, Tuner, AUX 1, AUX 2, Tape monitor, Mono, h.f. and l.f. filters. Plus switches for mains on/off and loudspeaker or phones.
The headphone socket is also mounted on the front panel. DIN sockets are fitted on the rear panel for all inputs, terminals for the loudspeaker connections, an earth terminal, and fuse for the mains and each loudspeaker output. The heat sinks for the power f.e.t.s are also mounted on the rear panel.

## Power f.e.t.s

Until recently bipolar transistors have been used in almost every Hi-Fi available (the exceptions using valves). Bipolar transistors require a wide area of Safe Operation to achieve reliability and a large gain-bandwidth product so that large amounts of negative feedback can be used at the higher audio frequencies to reduce the distortion. They also have a positive temperature coefficient which means that any increase in transistor temperature causes an increase in transistor collector current which causes the temperature to increase further, and so on requiring careful design to avoid thermal runaway. Also, as bipolar transistors are minority carrier devices, they suffer from storage effects at high frequencies which can cause a most objectionable distortion, which may well account for the so called "transistor sound".

In the light of all these problems a considerable amount of money and research has gone into looking for a better device. One result of this has been the power f.e.t.s developed by Hitachi Ltd., of Tokyo, Japan. Some advantages over conventional transistors are:

1) Good frequency response because of fast carrier speed.
2) No storage effect.
3) Negative temperature coefficients, so no risk of thermal runaway.
4) No secondary breakdown.
5) High input impedance and high gain.

The Winton amplifier uses a complementary pair of Hitachi power f.e.t.s Type 2SJ48 ( $p$-channel) and 2SK 133 ( $n$-channel). These devices have a maximum dissipation rating of 100 W each, so when used in an amplifier of 50 W output, each device is only dissipating about 25 W , or just ticking over! In fact combined with a 120 V drain to source breakdown voltage and a drain current of 7A they are almost indestructible when used in the Winton, which


[^3]


Fig. 1: Circuit diagram of the PW "'Winton'" stereo amplifier, showing Channel "a" (left) only. The following components are shared between the two channels: R12-15, 37, 38, 62-64; C2, 3, 8-11, 24, 40-45; Tr13, 14; IC1-3; D1, 2, 8-12; VR3; T1; F2; S9. The remainder are duplicated for the other channel

## specification

## Power Output

Continuous sine wave power, both channels driven
$50+50 W$ into $8 \Omega$

Power Bandwidth (power amp. only)
$-1 \mathrm{~dB} 15 \mathrm{~Hz}-100 \mathrm{kHz}$

Frequency Response (power amp. only)
$-0.5 \mathrm{~dB} \quad 10 \mathrm{~Hz}-40 \mathrm{kHz}$
$-3 \mathrm{~dB} \quad 5 \mathrm{~Hz}-150 \mathrm{kHz}$

Harmonic Distortion (50 + 50W)
$1 \mathrm{kHz} 0.013 \%$
$10 \mathrm{kHz} 0.015 \%$
$20 \mathrm{kHz} 0.018 \%$
$100 \mathrm{~Hz} 0.011 \%$
$20 \mathrm{~Hz} \mathrm{0.019} \mathrm{\%}$
No significant increase at lower powers

Intermodulation Distortion ( $28 \mathrm{~V} p k$ into $8 \Omega$ )
$\mathrm{f}_{1}=15 \mathrm{kHz} \quad 2 \mathrm{f}_{1}-\mathrm{f}_{2} 0.005 \%$
$\mathrm{f}_{2}=16 \mathrm{kHz} \quad 2 \mathrm{f}_{2}-\mathrm{f}_{1} 0.004 \%$
$f_{2}-f_{1} 0.003 \%$

## Damping Factor

$20 \mathrm{~Hz}-1 \mathrm{kHz} 80$
20 kHz 20

Rise Time (power amp. only) $2 \mu \mathrm{~s}$

Slew Rate (power amp. on/y) $26 \mathrm{~V} / \mu \mathrm{s}$

Stability - Unconditional

Sensitivity for 50W
Disc 3mV
Tuner 100 mV
AUX 100 mV
Tape 100 mV

Input Impedance
Disc $47 \mathrm{k} \Omega$
Tuner $100 \mathrm{k} \Omega$
AUX 100k $\Omega$
Tape $100 \mathrm{k} \Omega$

Maximum Input for $0.1 \%$ t.h.d.
Disc 140 mV
Tuner 4V
AUX 4V

Frequency Response $\pm 0.5 \mathrm{~dB}$
Disc RIAA
Tuner $20 \mathrm{~Hz}-20 \mathrm{kHz}$
$\mathrm{AUX} 20 \mathrm{~Hz}-20 \mathrm{kHz}$
Tape $10 \mathrm{~Hz}-40 \mathrm{kHz}$


Plot of Harmonic Distortion components generated when handling a 1 kHz signal dissipating 50 W in an $8 \Omega$ load. The 2nd harmonic is at -82 dB and the 3rd harmonic -86 dB relative to the fundamental


The upper trace shows the PW 'Winton'" output when handling a 1 kHz signal dissipating 50 W in an $8 \Omega$ load. Note the complete absence of crossover distortion

The lower trace shows the residual harmonic content of the above signal, after the fundamental had been removed by the distortion meter filter. The total harmonic distortion is about -78 dB , but is masked by noise


Plot of Intermodulation Distortion components generated when driven to full power into an $8 \Omega$ load by two equal-amplitude signals, $f_{1}=15 \mathrm{kHz}, f_{2}=$ 16 kHz . Component $2 \mathrm{f}_{1}-f_{2}$ is at $-\mathbf{8 6 d B}, 2 f_{2}-f_{1}$ at -88 dB , and $\mathrm{f}_{2}-\mathrm{f}_{1}$ estimated at about -90dB

This test, using these frequencies, is very critical of the power-bandwidth capability of an amplifier, and is seldom given in specifications due to the generally poor figures obtained

## Hum \& Noise

Unweighted* with reference to rated sensitivity
Disc-70dB
Tuner -75 dB
$\mathrm{A} U \mathrm{X}-75 \mathrm{~dB}$
Tape -75dB

## Hum \& Noise

Disc input with reference to 10 mV input Unweighted $80 \cdot 5 \mathrm{~dB}$

Cross Talk** -48 dB

Tape Output 100 mV
AUX 1100 mV via $100 \mathrm{k} \Omega$

* See comments regarding specifications.
** Important to note that cross talk residual is clean, i.e., does not introduce distortion into the other channel-a common failing in many amplifiers and seldom mentioned.


## Tone Controls

Bass $\pm 10 \mathrm{~dB}$ at 100 Hz
Treble $\pm 10 \mathrm{~dB}$ at 10 kHz

## Balance Control

+0.5 dB to zero, each channel

Channel Matching 0.5 dB

## Filters

| I.f. -3 dB 50 Hz | $12 \mathrm{~dB} /$ Octave |
| :--- | :--- |
| h.f. -3 dB 5 kHz | $12 \mathrm{~dB} /$ Octave |

Subsonic, disc only, non-switchable

| (IEC 65) -3 dB | 20 Hz |
| :---: | ---: |
| -8 dB | 10 Hz |
| -17 dB | 5 Hz |

Ultrasonic, all inputs except tape, non-switchable
$-3 \mathrm{~dB} \quad 60 \mathrm{kHz}$
$-7 \mathrm{~dB} \quad 100 \mathrm{kHz}$

Before making any measurements, the amplifier was pre-conditioned for 1 hour at 30\% full power with both channels driven. Measurements were made after a further 5 minutes at full power.
Unless otherwise stated, the volume control was at maximum, tone controls at centre and filters switched out.
The stability tests were made using various combinations of loads with capacitors up to $2 \mu \mathrm{~F}$ and also with capacitors up to $2 \mu \mathrm{~F}$ without additional load. Square wave signais were used over a frequency range of 20 Hz to 20 kHz .
The figures given are those obtained on the prototype amplifier.

## Test Equipment

Hewlett Packard HP3580 spectrum analyser; KronHite 4100 low distortion oscillator; Sound Technology 1700B distortion analyser, power output meter and low distortion oscillator; Telequipment D83 oscilloscope; Polaroid oscilloscope camera.
This equipment was kindly made available by Armstrong Audio Ltd., at their research laboratory.

## General Comment

Be very careful when comparing specifications of other amplifiers with the Winton. Many other published specifications use a weighted signal to noise figure. This would allow an extra 10 dB or so to be added to their figures and tends to favour the poorer quality amplifiers at the expense of the really good amplifiers.
means that the circuit can be kept simple as no protection circuits are required, eliminating another source of distortion.

## The Bi-f.e.t. Op.amps

The use of the Texas bi-f.e.t.s in the control unit section offers a number of advantages over the more usual 741 type of op.amp. The main ones being:

1) Wider bandwidth. 3 MHz typical.
2) High slew rate $13 \mathrm{~V} / \mu \mathrm{s}$.
3) Low distortion $0.01 \%$.
4) High input impedance j.f.e.t. input stage.
5) Low noise $18 \mathrm{nV} / \mathrm{Hz}$ (TL072CP).
6) 80 dB supply ripple rejection.

When used in low or unity gain circuits using large amounts of negative feedback the noise and distortion from the device is almost unmeasurable. The Winton amplifier has a basic sensitivity at the AUX inputs of 100 mV and this has to be increased to 4 V before the distortion reaches $0 \cdot 1 \%$. At any normally used input level the distortion is below the noise and completely inaudible.

## Circuit Description

Both channels are identical and are pre-fixed "a" for left and " $b$ " for right. Only the "a" channel will be described. See Fig. 1.

## The Disc Pre-Amplifier

The input from a magnetic cartridge is fed into the base of $\operatorname{Tr} 1$, a low noise BC 414 via the input load $\mathrm{R} 1,47 \mathrm{k} \Omega$, and an r.f. filter R2C1. The transistor $\operatorname{Tr} 1$ forms one half of a differential input pair. Operating at a low current, approximately $100 \mu \mathrm{~A}$ each half, ensures the minimum amount of noise.

The output from $\operatorname{Tr} 1$ is coupled directly to $\operatorname{Tr} 3 \mathrm{a}$ BC556. This in turn is coupled to a subsonic filter (as recommended by IEC65) consisting of C12, R16 and R17. This subsonic filter has its -3 dB point at 20 Hz , Fig. 2. This response, coupled with the normal RIAA response provides a 12 dB /octave filter at subsonic frequencies and prevents intermodulation at low frequencies caused by warped records, etc., from being produced and affecting the overall sound quality. The RIAA negative feedback equalising circuit consists of R7, R8, C $5, \mathrm{C} 7$ and is connected between the output of $\operatorname{Tr} 3$ and the base of $\operatorname{Tr} 2$.

In order to ensure further the low noise factor in the disc amplifier, electronic decoupling is used, $\operatorname{Tr} 13$ (BC546) Tr14 (BC556) in each supply rail to remove any ripple or noise from the power supply. This supply has already been decoupled and stabilised by the Zener diodes, D1, D2.

The output from the disc pre-amplifier then goes to the selector switches S 1-5.

## Control and Filter Section

From the selector switches S 1-5, each channel goes to one half of ICl a bi-f.e.t. op.amp, Type TL072CP. This is used as a low noise buffer amplifier (with 4 dB of gain) to provide a high impedance input, and a low impedance output suitable for driving the next stage, which also uses a bi-f.e.t. op.amp, Type TL082CP. This is used as an active high- (C15, C16, R22, R23) and low- (C17, C18, R24, R 25 ) pass filter. Operating at unity gain, the filter is designed to provide a $12 \mathrm{~dB} /$ octave cut-off with the -3 dB points at 50 Hz and 5 kHz respectively, Fig 3 . A slope of $12 \mathrm{~dB} /$ octave is considered to be optimum for Hi-Fi use as a steeper slope could introduce ringing on transients, which would sound most objectionable.

The tone controls follow the filters, again using a bif.e.t. op.amp, Type TL082CP, and the circuit is a Baxandall negative feedback type.

This type of tone control circuit keeps distortion and noise to a much lower level than the passive type and also provides frequency response contours more acceptable to the ear, Fig. 4.

The amount of bass and treble boost and cut has been restricted to around + or -10 dB at 100 Hz and 10 kHz . Although it is possible to obtain up to 20 dB boost and cut at these frequencies by removing R29 and reducing R33, extreme amounts of boost or cut are considered by the author to be bad design, as in practice large amounts can rarely be used. For example, suppose we have a signal requiring +10 dB at 100 Hz , with our circuit we can obtain this with only an extra dB or so at lower frequencies. If the unrestricted circuit was used, 10 dB of boost at 100 Hz would also produce 20 dB of boost at 20 Hz , as the response would continue to rise as shown in Fig. 4. The effect of this would be to increase all the rumble and other low frequencies a further 10 dB over the required level at 100 Hz . This would be in effect asking the amplifier for a considerable increase of power at low frequencies and as it cannot provide this, severe distortion would result. Note:

## 10 dB of boost equals a voltage ratio of $3 \cdot 16: 1$.

If our amplifier is already producing 50 W into $8 \Omega$, i.e., 20 V , then $20 \mathrm{~V} \times 3 \cdot 16=63 \cdot 2 \mathrm{~V}$, power $=\mathrm{V}^{2} \div \mathrm{R}=3994 \div$ $8=499 \cdot 28$ watts!!

Even with the amount of boost restricted to +10 dB , if the bass control is at maximum and the amplifier is just reaching maximum power (50W) at say 20 Hz then the maximum power at 1 kHz must be limited to 5 W (assuming a flat frequency response of the input signal). In practice, the tone controls are normally used to make up for the deficiency in the incoming signals, i.e., to restore the signal to an overall "flat" response, so the power restriction normally won't apply.

The output from the tone control circuit then passes to the balance control and the tape monitor switch S8. This switch selects either the output from the control unit section or the output,from a tape recorder (connected to the tape socket) and feeds the signal via the volume control to the power amplifier. A tape output signal is permanently connected from the control unit output to both the tape socket and via $100 \mathrm{k} \Omega$ resistors to the AUX 1 socket. This arrangement allows the use of either or both reel to reel and cassette recorders with the Winton. When using a reel to reel or cassette recorder fitted with a monitor head (plugged into the "tape" socket) tape monitoring can be achieved by simply pressing the "tape" button.


Fig. 2: Disc input - Subsonic filter response

Fig. 3: Frequency response of l.f. and h.f. filters

Fig. 4: Frequency response range of Bass and Treble controls

Fig. 5: Cross-talk between the two channels, for Disc and Auxiliary inputs

## Power Amplifier

The circuit used for the power amplifier is based on a design produced by Hitachi for their power m.o.s.f.e.t.s. The original circuit has been modified by the author to suit the requirements of the Winton and electronic decoupling added to further improve the overall specification.

The signal from the volume control, VR4 is fed via an r.f. filter R40, C27 to the base of $\operatorname{Tr} 4$. This is one half of a differential pair $\operatorname{Tr} 4, \operatorname{Tr} 5$ using low noise transistors Type BC556. These in turn drive a second differential pair, Tr6, Tr7, Type 2SC1775, which has an active collector load (current mirror formed by $\operatorname{Tr} 8$ and D3), to maintain the pushpull action. The power m.o.s.f.e.t.s $\operatorname{Tr} 11, \operatorname{Tr} 12$ are driven directly from the second differential pair. As the



inherent distortion of the amplifier is very low, only 45 dB of negative feedback is used, this is fed from the output, back to the base of $\operatorname{Tr} 5$ via R46, R45, C31. The use of less feedback means that, at the overload point, the onset of distortion is less severe then would be the case with amplifiers using considerably more negative feedback. Diodes D4, D7 and Zeners D5, D6 form an overdrive protection circuit to prevent the gates of the m.o.s.f.e.t.s from receiving excessive drive voltages in the event of a fault condition.

Due to the excellent high frequency response of the amplifier only minimal feedback phase correction is required via C31, R45, and L1, R58, R59, C39. The amplifier is unconditionally stable.

As mentioned earlier, electronic decoupling has been included in each power amplifier to reduce the power

## components




Internal view of the PW "'Winton'' amplifier. Here, we show the final version
using a printed circuit board. The model illustrated on our front cover is the
prototype, which was constructed on perforated board with hard wiring
supply noise and ripple voltage, as well as improving the cross-talk (Fig. 5) between channels. Transistors Tr9, BC546 and $\operatorname{Tr} 10$, BC556 are used for the decoupling circuit.

Provision is made to adjust the d.c. offset at the output to zero. By adjusting VR5, the current through the differential input stage can be adjusted to compensate for the slight gain variation in each half of the differential pairs. It is very important that the output from the amplifier does not contain any d.c. voltage, as this would cause the loudspeaker cone to take up a position away from its centre which in turn would cause distortion because of the resulting non-linear cone movement.

## Power Supply

By no means least important, the power supply of the Winton has been subjected to the same standard of design as the rest of the circuits, although on the surface it may look simple.

In a high fidelity amplifier the power supply can either turn a good amplifier into a top quality model or into a not so good one. There are two main reasons for this:

1) The regulation of the supply volts must enable the amplifier to reach its full power potential under the worst possible signal conditions. At the same time, it must not let the voltage rise to dangerous levels under quiescent conditions.
2) The external magnetic field from the transformer must not inject any hum into the amplifier circuits, particularly the disc amplifier.
To enable the full performance that the Winton is capable of to be achieved, it was decided to use a toroidal type of transformer. The design and development of this special low-field transformer was carried out by Belclere Ltd., and this transformer is now available from T \& T Electronics.

The output from the transformer goes to a bridge rectifier and then to the two $4700 \mu \mathrm{~F}$ capacitors C40, C41. These, and the electronic decoupling circuits used in the
power amplifiers, coupled with the Zener stabilised supply to the control unit and further electronic decoupling to the disc pre-amplifier, ensure that only about $0.001 \%$ of the total output from the Winton is noise from the power supply. Even on the disc input the total noise in the output from all sources combined is only about $0.04 \%$.

## Heat Sink Ratings

When designing an amplifier for domestic use, the design is almost always a compromise between all the various requirements. In the case of the Winton amplifier this applies to the size of the heat sink and mains transformer.

It is possible to make both of these components in such a way that the amplifier could deliver $50+50 \mathrm{~W}$ continuous power for hours on end without much temperature rise. The cost of doing this would be so high as to put the amplifier out of reach of most peoples' budget. On the other hand these components could be made down to a low price and the amplifier allowed to run very hot, which would lower the reliability.

In the design of the Winton the author has arranged that these two components will allow the power f.e.t.s and the transformer to run well inside their respective maximum rating under continuous drive conditions, although under these conditions both the heat sink and transformer will reach quite high temperatures and certainly be too hot to touch.

When used for its normal purpose, i.e., reproducing music, the temperature rise is very much less and after some hours of loud music will still be only moderately "Hot". The actual heat sink may reach $60^{\circ} \mathrm{C}$ under these conditions.

This compromise doesn't reduce the quality of the reproduced sound in any way, but it does avoid a severe pain in the wallet.

## NEXT MONTH Constructional details

Those who built the Avon transmitter described in $P W$, July 1978 , may be interested in one or two simple yet significant modifications which were ultimately incorporated into the prototype. The purpose was to increase the power output to approximately 20 watts for the original 24 volt supply and to give some consideration to a repeater access facility. The question of powering the transmitter from 12 volts-a car battery, for instancewas also examined, and the update caters for this at a reduced r.f. output of around 9 watts.

In the first piace, the "Calibrate" facility provided by S1 in the switching arrangements of Fig. 10, p. 52 ( $P W$, Aug. '78) was dispensed with and the switch re-assigned to a toneburst facility for repeater working. Details of the toneburst module appear elsewhere in this issue: the board on which it is constructed can easily be accommodated in the availabie space.

Step two is to replace the 15 V regulator (7815) with a 12 V type 7812 . Our attentions are then directed towards the power amplifier board, where the majority of changes take place. Here a little care is needed, although the modifications are far from complex. The revised board could, of course, be constructed as an "afterburner" which would provide a substantial increase in output from, say, a hand-held tranceiver.

Initially, remove the resistor R1 ( $82 \Omega 2 \mathrm{~W}$ ) and take the h.t. feed (now 12 V ) directly to the board. The collector supply to the p.a. transistors remains routed to the input of the 7812 voltage regulator, via the switching of Fig. 10 (Aug.' 78 ).

Now take out the BLY83 transistors and carefully expand the holes in the p.c.b. to accommodate the new 2N5642 devices. The leads will require a little trimming and this should be done fairly accurately to permit the transistors to fit on the appropriate islands on the board.

The increased output of the new p.a. inevitably means that appreciably more heat will be dissipated by the final stages. Additional heatsinking will have to be provided and in the prototype this took the form of a piece of aluminium $100 \times 50 \times 4 \mathrm{~mm}$, drilled to accept the studs of the power transistors. The sink is fixed to the main chassis by means of a suitable nut and bolt in each corner and the surface in contact with the chassis coated with Thermopath grease or similar compound. Insert the 2N5642 devices and tighten down, carefully soldering the connections to the p.c.b. afterwards (Fig. 1).


Fig. 1: Mounting the power transistors

The installation of the toneburst module designed by Philip Hodson is extremely simple. A power supply is taken, via the defeat switch S1, from across the energising coil of the keying relay RLA, which is shunted with a diode to absorb switching transients. Miniature screened lead feeds the signal input to the base of $\operatorname{Tr} 2$ (Board 1), with a $1 \mu \mathrm{~F}$ tantalum capacitor placed in series.

The prototype exhibited some problems which ultimately were attributed to stray r.f. entering the microphone input. It is therefore recommended that a


Fig. 2: The new switching arrangements - note the earth on changeover contacts RLA2, the option of different coil resistances for RLA and RLB from the original article, and the addition of diodes D2 and D3

# Wideband RF Noise Source <br> <br> D.Whitfield 

 <br> <br> D.Whitfield}

A wideband r.f. noise generator is a most useful device for carrying out rapid performance checks on all types of h.f. and v.h.f. receiver. The wide variety of applications of such an instrument includes gain checks, receiver sensitivity measurements and, where a calibrated attenuator is available, the noise generator may also be used to carry out accurate measurements of gain, noise figure, a.g.c. characteristics, and the calibration of signal strength meters.

The noise generator described here uses a minimum of components, is simple to construct and, although its output level is uncalibrated, is eminently suitable as a source for receiver alignment with an output which is substantially of constant level over a very wide frequency range. A previous article dealt with the construction of a matching calibrated $50 \Omega$ attenuator.

True "white" noise can be thought of as a signal which is evenly distributed over an infinitely wide range of frequencies and, although the instantaneous voltage varies randomly, the r.m.s. noise voltage developed into a resistive load will be constant when measured over a short time period. The noise power developed into a purely resistive load is thus directly proportional to the bandwidth in which it is measured.

## Generating Noise

The simplest form of noise source is that due to thermal agitation current in a conventional carbon resistor. The fluctuation in current caused by the random movement of electrons is evenly distributed in frequency but the noise voltage produced is usually extremely small. Practical noise generators do not provide an infinite noise spectrum. This is due both to the limitations of the noise source itself and to the bandwidth of any following amplifier. The choice of noise source depends on a number of factors.


Fig. 1: Block diagram (a) and Thevenin equivalent circuit (b) of the noise source


Thermionic diodes are popular in commercial noise generators but, for economy of supply and portability, the noise source employed in this design is a conventional Zener diode. A considerable noise voltage is developed across the series load resistor of a Zener diode when it is undergoing breakdown. It is, however, usually necessary to include a stage of amplification following a noise source so that its output voltage is of a sufficient level for general purpose use. This has the added advantage that the amplifier stage also provides a degree of isolation between the noise source and the input of the circuit or receiver under test.

## Circuit Description

A block diagram of the noise generator, together with its Thevenin equivalent circuit, is shown in Fig. 1. The noise output of the Zener diode is amplified using a twostage wideband amplifier, $\operatorname{Tr} 1$ and $\operatorname{Tr} 2$. Both transistors operate in the common emitter mode and the gain of the second stage, $\operatorname{Tr} 2$, is made variable by the application of negative feedback by means of VR1. The device used in the $\operatorname{Tr} 2$ position has an effect on the bandwidth of the noise output. For use at v.h.f. the transistor recommended is a BFY 90 as this has a very high cut-off frequency. For general purpose use at h.f. (i.e. to at least 30 MHz ) a much less expensive transistor may be used, and the 2 N 706 is quite adequate.

In order to facilitate matching to a standard $50 \Omega$ system, the output of the wideband amplifier is terminated in a $56 \Omega$ carbon resistor. This, together with the parallel effect of the collector load resistor of Tr2, ensures an output resistance of almost exactly $50 \Omega$. The output noise voltage can be thought of as being generated in this resistor and thus the Thevenin equivalent circuit of Fig. 1b is realised.


Fig. 2: Circuit diagram of r.f. noise source

## Construction

Construction is straightforward and the use of the printed circuit board layout described is highly recommended. Where a printed circuit board is not used, care must be taken to ensure that all wiring is neat and as direct as possible. The noise generator is housed in a standard diecast box. Any alternative screened metal box can be used provided that the leads from the printed circuit board to the output socket are kept as short as possible. A small


Fig. 3: Internal arrangement of components

## $\star$ components





Fig. 4: Printed circuit board copper track pattern shown full size

Fig. 5: Component placement
on p.c.b.

hole is made in the front panel of the box to provide access for the adjustment of VR1. Either a skeleton type or a knob type pre-set may be used. The latter will prove to be more satisfactory where frequent adjustments of output level are likely to be made. Care should be taken to ensure that the hole made in the case aligns correctly with the adjusting slot or shaft of the pre-set. Standard variable resistors are not recommended for use since they are not only too large but have considerable stray reactance which may impair the frequency response of the amplifier. The BFY90 is available in both TO18 and TO72 packagesthe latter type should have pin 4 (case connection) clipped off before soldering the remaining connections to the board.

## Testing

When the wiring of the noise generator has been completed, a thorough visual check should be made before connecting the battery. Connect the noise generator to the input of a receiver and check that an output is obtained. If the receiver has a signal strength meter this should show a steady indication; if no signal strength meter is fitted a 'rushing' sound should be heard from the loudspeaker. Adjustment of VR1 should cause a change in the noise level. If possible, a check should be made over a range of frequencies. If the receiver sensitivity is constant over a range of frequencies its signal strength reading (or audible output level with a.g.c. off) should remain constant as the receiver is tuned across its entire frequency coverage. Receiver alignment may be carried out using the receiver's signal strength meter in a conventional manner. The use of an a.f. voltmeter connected across the loudspeaker terminals as an output level indicator is strongly recommended when alignment is carried out on receivers not fitted with a signal strength meter.


Amongst the majority of would-be and newly-licensed amateurs there still seems to exist an air of uncertainty with regard to repeaters. This is quite apparent by the number of new 2 m stations openly admitting "I'm recently licensed and don't fully understand repeater operation". This article looks at the principle of repeater working and suggests a suitable design for a toneburst generator with which to access them.

## Operation via Repeaters

A repeater may be simply described as a remotelyoperated transceiver which simultaneously transmits on one frequency and receives on another. Such a function is usually referred to as duplex, as opposed to simplex, when transmission and reception occur on the same channel but not at the same time. In the duplex mode, the repeater transmit and receive frequencies are well separated, the transmit channel being the higher of the two. This difference is referred to as the "shift" and is 600 kHz in the case of 2 m and 1600 kHz at 70 cms .

The amateur therefore would normally transmit on the repeater receive channel, which is known as the input. Some confusion arises here, especially when the term "reverse repeater" is used. This, as the name implies, refers to the amateur transmitting on the repeater transmit channel, effectively by-passing the facility. The function can be useful when calling a station known to be within point-topoint distance, and under normal circumstances you would transfer to one of the simplex channels when contact has been established. For obvious reasons you should not call a station in this mode unless you can hear him on the repeater input.

Table 1: 2m Repeater Channels

| Channel | Frequency <br> (Receive) | Frequency <br> (Transmit) |
| :---: | :---: | :---: |
| R0 | 145.000 | 145.600 |
| R1 | 145.025 | 145.625 |
| R2 | 145.050 | 145.650 |
| R3 | 145.075 | 145.675 |
| R4 | 145.100 | 145.700 |
| R5 | 145.125 | 145.725 |
| R6 | 145.150 | 145.750 |
| R7 | 145.175 | 145.775 |
| R8 | 145.200 | 145.800 |
| R9 | 145.225 | 145.825 |

Table 1 shows the frequency and channel allocations for duplex repeater operation. Not all of these are at present in use within the UK however, our own stations being between R3 and R7 inclusive.

Access to a repeater is gained by a short burst of tone $(1750 \mathrm{~Hz})$ generated at the beginning of the transmission and lasting for 500 ms typically. Some repeaters require this to be followed by several seconds of speech without loss of carrier, and this is quite difficult to arrange if all the switching is done manually. The device described in this article fulfils all these functions by one operation of the push-to-talk switch.

Most repeaters, with the notable exception of GB3LO (London), restrict the duration of transmission. After a pre-determined period a time-out occurs and the repeater releases, having first sent a series of "pips" to advise the receiving station of its intention.

When the input of the repeater senses the absence of a carrier, an advisory signal is sent, indicating that the station is ready to be re-accessed. Here it should be clearly understood that the terms "access" and "re-access" do not relate to the same function. Re-access involves the resetting of the timing mechanism immediately after the repeater has been used, whereas access is given to mean "starting from cold", as it were.

British repeaters vary considerably in the method by which re-access takes place. Some require a toneburst with or without speech to follow, others operate on receipt of speech alone whilst some require only to receive a carrier, whether modulated or not. Some very interesting information is given in the International VHF-FM Guide $\dagger$, for those who would like to consider individual cases.

## The Toneburst Module

Referring to the circuit of Fig. 1, we can see that the crystal oscillator operates at a frequency of 910 kHz . The Schmitt trigger IC 2a converts the waveform into a sharpedged square wave which is fed to the twelve-stage ripple counter IC3 as the clock reference. This counter, in conjunction with other gates, divides the clock frequency by 520 , thereby obtaining the required 1750 Hz necessary. The output is routed via another Schmitt trigger IC2c into a 3-pole Butterworth filter, which produces a relatively pure sine wave.

[^4]

Immediately power is applied, the capacitor C9 starts to charge through R10. As the potential rises towards that of the supply, a point is reached where IC 2c changes state, effectively blocking the path of the 1750 Hz signal. When this occurs, the system has to be re-set by disconnecting the power supply and consequently allowing the capacitor to discharge.

By installing the toneburst module in such a way that on pressing the microphone p.t.t. switch power is simultaneously applied to the transmitter, the access signal will be generated on the initiation of each "over" and the possibility of dropping carrier is eliminated. The provision of a switch in the power supply to the module permits its isolation when working simplex.

## Construction

The discrete components should be mounted initially, leaving the crystal and integrated circuits until last. Omit resistor R12 at this stage, otherwise it will have to be removed in order to test the board.

If you do not wish to use holders, leave IC 1-3 until last. These devices are c.m.o.s. and so should be handled as infrequently as possible prior to installation. The techniques for soldering these components have been considered on many occasions within these columns and in other publications. These should be closely adhered to if damage is to be avoided from static charges. The crystal is also susceptible to damage from excessive heat. The best method of soldering this is to use a very hot iron and to complete the process as quickly as possible.

## components

|  |  |
| :---: | :---: |



Fig. 2 (left): Copper track layout of the p.c.b. shown full size and (right) the component layout

The capacitor C9 is specially selected, and it has been found that only the type in the components list is suitable. Most tantalum beads have a tolerance of $20 \%$, and these gave rise to problems in the prototypes. For this reason a tubular version with axial leads and a tolerance of $10 \%$ is required. Increasing the value of this device will lengthen the duration of the toneburst.

## Testing the Board

Make certain that you have not fitted R12 before commencing. Now connect a small loudspeaker or pair of headphones between the slider of VR1 and earth, then temporarily short across C9 with a small piece of wire. Apply around 15 volts and you should hear a continuous 1750 Hz tone. Disconnect, and remove the short across C9. Re-apply power, and the tone should occur for a period of approximately half a second. Should it happen that the burst is appreciably longer, then reduce the value of the capacitor slightly. Selection of the correct value proved to be the only variable in the prototypes. The final decision to use $4.7 \mu \mathrm{~F}$ was taken only after exhaustive experimentation, which indicated that this value produced the most consistent results. Provided the burst duration falls between 540 and 650 ms , no difficulties should be encountered.

After carrying out these procedures, don't forget to insert R12!

If you are installing the module into the Avon transmitter, the following information may be helpful. Connect a microphone to the transmitter, key the device and adjust VR1 on Avon Board One to obtain a clear and undistorted speech output. Do not attempt to set for maximum volume; this is not the function of the deviation control.


Fig. 3(a): The toneburst module fitted to a simple push-to-talk arrangement


Fig. 3(b): Fitting the module into a high-power transmitter, where a keying relay is used

Too little deviation is preferable to an excess. Once adjusted, do not touch this control again, but alter VR1 on the toneburst module to give a level of tone slightly lower than that obtained with speech. If you are lucky enough to have access to an oscilloscope, the speech should be set for a maximum deviation of 5 kHz and the toneburst limited to 3.5 kHz .

## Follow-up to the PW Avon

continued from page 48
100 pF silver mica capacitor be connected from the input to earth, as close to the p.c.b. as possible, to provide the necessary decoupling.

Replace the p.a. board and the modifications are now complete. Retuned, the transmitter will produce about 9 watts of r.f. with a power supply of 12 volts, rising to around 20 watts for the original 24 volt supply.

Operating the microphone pressel switch with the "Access Repeater" facility enabled will automatically generate toneburst, thus providing the required conditions to open repeater stations.

The power supply featured in the September ' 78 issue of $P W$ can easily be modified to handle the increased current drain, provided the transformer is capable of delivering 1 it . Merely place another $0.47 \Omega 25$ watt precision wirewound resistor in parallel with R2 (Fig. 1, p. 30, September '78 issue).

## Q D D D TONEBURSTKIT ASINTHISISSUE

| 910 KhzXtal | 2.16 | 4093 |
| :--- | :---: | :--- |
| 3.3 FTant | 27 p | 4040 |

When using a small current (such as that from a t.t.l. or c.m.o.s. logic device) to control a much larger current in a load, it is possible to employ a simple transistor amplifier in some cases, whereas if a higher current gain is required a power Darlington device may be used as an amplifier. However, simple circuits employing such devices do not incorporate any means of limiting the current in the load so as to protect both the load and the amplifier device from possible damage. Also they do not incorporate any means of protecting the amplifier device if it should undergo an excessive temperature rise.

The TDE 1607 is a new integrated circuit from Thomson-CSF which can be used to overcome these problems. Basically it is a protected operational amplifier, being designed for fairly high currents and voltages, and is specifically intended for the control of the current in such loads as relays, lamps and stepping motors.

The TDE 1607 is essentially fail-proof in operation provided that the absolute maximum permissible supply voltage and input voltage (both 36 V ) are not exceeded. An external resistor can be incorporated in the circuit so as to limit the current in the load to any chosen value up to 0.5 A (although the TDE 1607 has a maximum permissible output current of 1 A ). If the silicon chip in the device becomes too hot for safety, the internal circuit will automatically reduce the output current and thus prevent possible damage.

## Connections

The TDE 1607 is supplied in a small circular metal transistor-type package with 6 leads. The connections are


Fig. 1: Package outline and connections


Fig. 2: Block diagram and 'pass' transistors
shown in Fig. 1, but it must be remembered that this is a top view. This device can operate over the wide temperature range of $-25^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$.

The way in which the device operates can be understood using the block diagram of Fig. 2. When the non-inverting input to the amplifier (marked + ) is at a voltage greater than that of the inverting input at pin 2 (marked -), the output voltage from the amplifier at point A in Fig. 2 will be relatively high (only a little below that of the $\mathrm{V}^{+}$supply line).

The output from the amplifier will then normally feed a current into the base of the transistor $\operatorname{Tr} 2$ (inside the TDE 1607) so that this transistor conducts and passes a current through the external resistor $\mathrm{R}_{\mathrm{sc}}$ and through the load, $\mathrm{R}_{\mathrm{t}}$. The value of the current sensing resistor $\mathrm{R}_{\mathrm{sc}}$ determines the maximum current which can flow in the output circuit. For example, if the value of $\mathrm{R}_{\mathrm{sc}}$ is about $1 \cdot 5 \Omega$, a current of 0.5 A will produce about 0.75 V across $\mathrm{R}_{\mathrm{sc}}$ and this voltage is applied between the base and emitter of the transistor Trl. It is adequate to cause this transistor to commence to conduct, so instead of the current from A passing to the base of $\operatorname{Tr} 2$, it will pass to the collector of Tr1. Thus $\operatorname{Tr} 2$ will pass only just enough current to maintain a large enough voltage across $\mathrm{R}_{\text {sc }}$ to keep $\operatorname{Tr} 1$ conducting. The current passing through $\mathrm{R}_{\text {sc }}$ is equal to the load current, so the internal transistors Tr 1 and Tr 2 can be used to limit the load current.

If $\mathrm{R}_{\text {sc }}$ has the value of 1.5 ohms mentioned above, the load current will be limited to about 0.5 A , whereas as $\mathrm{R}_{s c}$ is increased to 3 ohms, the maximum load current will be
about 0.25 A . The writer found experimentally that still larger values of $\mathrm{R}_{\text {sc }}$ could be used, a value of $100 \Omega$ producing a maximum current of some 10 mA . The exact value of the maximum current varies somewhat with the case temperature of the device, typical values being plotted in Fig. 3. Obviously the output current will never exceed the value of $\mathrm{V}^{+}$divided by $\left(\mathrm{R}_{\mathrm{L}}+\mathrm{R}_{\mathrm{sc}}\right)$, since an adequate voltage is required to cause the current to flow.
The output current will flow only when the noninverting input of the internal amplifier (pin 3) has a potential exceeding that of the inverting input (pin 2); if this condition is not satisfied, the device will be 'off' and little output current will flow. If the silicon chip of the device ever becomes too hot for safety, the thermal protection circuit of Fig. 2 passes current from point A to ground so that there is little bias current to feed the internal transistor $\operatorname{Tr} 2$; the output current therefore falls until the device cools. However, it is unwise to operate the device for an appreciable time at such a high temperature that the thermal protection circuit is in operation, since such high temperature operation can tend to impair device reliability and certainly tends to produce surface defects on the chip. It should rather be regarded as a safety circuit which is not normally used.


Fig. 3: Typical output current plotted against case temperature

## Power Supply

It is recommended that TDE 1607 device is fed from a power supply of between 10 V and 30 V ; it is, of course, possible to use $\pm 15 \mathrm{~V}$ balanced supplies if desired. The supply current is typically only 3 mA (maximum 5 mA for any TDE 1607 device) from a 24 V supply, when the output current is zero.


Fig. 4: Graph showing safe operating area

When the output voltage from the device is low, ideally no output current should flow and the output voltage should be zero. In practice, an output leakage current of up to $100 \mu \mathrm{~A}$ can flow through the load at a junction temperature of $20^{\circ} \mathrm{C}$ or up to $500 \mu \mathrm{~A}$ at a temperature of $85^{\circ} \mathrm{C}$, whilst the output voltage may be a volt or so above the potential of the ground line. Similarly, when the output voltage is in the high state, there may be a potential of up to 1.8 V between the output and the positive supply line.

The input currents required by the TDE 1067 are quite small, typically 100 nA (maximum $1 \cdot 5 \mu \mathrm{~A}$ ). The input offset voltage at which switching from one state to the other occurs is typically 2 mV (maximum 50 mV ) between the two inputs.


Fig. 5: An application circuit of the device

In order to provide adequate protection for the device, the value of the resistor $\mathrm{R}_{\mathrm{SC}}$ should be chosen so that one operates within the unshaded area of the graph of Fig. 4. For example, with supply volts of up to 24 V the value of $\mathrm{R}_{\mathrm{sc}}$ should not be less than $0 \cdot 8 \Omega$ so that the output current is never allowed to exceed about 1 A . Obviously there is no objection to the use of larger values of $\mathrm{R}_{\mathrm{sc}}$ where a smaller output current is adequate for the application concerned.

## Typical Circuits

A typical application circuit for the TDE 1067 is shown in Fig. 5. The inverting input is biased by the potential divider to the desired value; a load current flows when the potential of pin 3 exceeds that of pin 2.

The diode D1 is required only if the load is inductive, such as a relay. When the current ceases to flow through the relay or other inductive load, a short transient reverse voltage can be developed across the relay coil. The diode is used to short this transient voltage to ground, since transients can damage the TDE 1067 device.


Fig. 6: Extension of the Fig. 5 circuit providing higher output current

## 5A Output

To provide a higher current than the TDE 1067 can itself supply, the circuit of Fig. 6 may be used in which an extra pnp power transistor fitted to a heat sink is used with the TDE 1067. Current limiting is again provided by the resistor $R_{s c}$, but the value of this component is much smaller owing to the higher current passing through it. Such small resistors are conveniently made from a length of resistance wire wound around a high wattage resistor of much higher value.

## Conclusion

The TDE 1067 is a versatile device which can be operated by very low currents to its inputs. When used correctly in circuits, it is almost impossible to damage the device even if the output is short-circuited to ground. At the time of writing the price is $£ 2.99$ (including VAT) plus $£ 0 \cdot 20$ for packing and postage from Phoenix Electronics Ltd, 46 Osborne Road, Southsea, Hants PO5 3LT.

## RIDIW IOIE:

Car Radio LW Converter, December 1978.
The MVAM 15 varicap diodes are sfown on the circuit diagram but omitted from the components list. They are avallable from Ambit Intemational, Gresham Rd. Brentwood, Essex.

Breadboards Supplement; December 1978.
On the 5 V regulated supply (p. 2 of supptement), DI is shown incorrectly. The configuration should be identical to that of D2.

Digital Door Chimes, December 1978.
tin the audio amplifier section in Fig. 1. Tr3 (BC160) is shown invented. The emitters of Tr 2 and Tr 3 should be linked. The collector of Tr3 goes to chassis.

Sandbanks Metal Detector. January 1979.
C17, 18, 19, and 20 are incorrectly shown in the components list ( p .50 ) as 10 nF , These should all be $\ln \mathrm{F}$, as correctly shown on the circuit diagram on p. 49. In the circuit diagram; Fig. 2 , diode 3 should be reversed in polarity.

## PW "Purbeck" Oscilloscope

Some readers seem to have misunderstood the information given in the Follow-up article $1 p, 55$ January 1979 ) regarding the mains transformer.

Due to an unfortunate combination of circumstances, the differing chatacteristics of the transformers supplied by Watford Electronics with their PW "Purbeck" kits were not picked up by the author when incorporating the sample transformer in the original PW "Purbeek" oscilloscope This oscilloscope and transformer have been in daily use now for many months, but two changes have proved necessary as follows.

If you are using one of the original Watford Electronics transformers then, as stated in the follow-up article, alt that need be done is to replace R103 by a wire shorting link, and to reduce the value of 101 to around $120 \Omega(5 \mathrm{~W}$ wire-wound; or better still, replace R101 by a wire shorting link. This provides a greater margin for low mains voltage It is NOT necessary to change the transformer:

Don't forget to check, as stated previousiv, that the $X$ and $Y$ Boards draw 20 mA and 30 mA respectively from the $\pm 150 \mathrm{~V}$ stabilised supply.

Please note that, when adjusting the input attenuator trimmers (C3-C10), a metal-bladed screwdriver is not suttable. An insulated coll-core adjuster should be used, and the hand held well clear to reduce stray tapacitance effects.

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Glasgow：Bo west Regent Street，G2 2CDD．Tel：04：－232 4133，Bristol： 1 Straits Parade，Fistponds Foad，B516 2t X，Tef：0272 654201．

| 2N929 | 0.37 | 2N3417 | 0.25 | 2N4062 |  | 2N5245 | 0.37 | AF106 | 0.80 | BC182L | 0.15 | AY－3－8500 | 6.50 | CA30 | 1.83 | LM34015 | 0.03 | LM | 70 | M7805KC | 15 | SN76013N | 50 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2 N 930 | 0.37 | 2 N 3439 | 0－35 | 2 N 4121 | 0.27 | 2N5248 | $\underline{+14}$ | AF109 | 0 | BC183A | 17 | CA3000 | 3.30 | CA3052 | 1.78 | LM340T12 | 0.83 | LM741C8 | 0.30 | LM7812KC | 1.75 | SN76012NO | 1.30 |
| 2 N 1131 | 0.32. | 2N3441 | 0.92 | 2N4122 | 4.27 | 2N5293 | 4.4 | AF114 | 0.78 | BC183LA | － 15 | CA300 | 4.25 | CA3053 | 0.77 | LM340115 | 0.83 | LM 741 C74 | 0.30 | ［M7815КС | 1.75 | SN76018KE | 60 |
| 2N1303 | 0.30 | 2N3442 | 1.45 | 2 N 4123 | 089 | ${ }^{2}$ N5294 | 14 | 4F． 15 | 0.7 | AC1 84 | 4.12 | CA3002 | 3.30 | CA3054 | 4.10 | LM340124 | 0.83 | LM747CN | G．99 | LM7824kC | 1.75 | SN76023N | 56 |
| ${ }^{2} \mathrm{~N} 1305$ | 0.30 | 2N3565 | 0.25 | 2N4124 | 419 | 2N5401 | 44 | Fint | 0.15 | EC184L | 0.15 | CA3005 | 2.50 | CA3059 | 2.16 | LM341P5 | 0.80 | LM748－8 |  | LM78205C2 | 0.30 |  |  |
| 2N1501 | ［ 30 | 143566 | $0 \cdot 25$ | こN4125 | －18 | ${ }^{2} \mathbf{N} 5415$ | 1．65 | ${ }^{\mathbf{c}} \mathbf{7}$ | 0.74 | BC205 | $9 \cdot 12$ |  |  |  |  |  |  | LM748－8 | 0.50 0.50 | LM78L12C2 | 0.30 | SN76023N0 | 1.36 2.35 |
| 2N1613 | 0.30 | 2K3567 | 0.25 | － CN 4125 | ${ }^{1} 18$ | ${ }^{2} \times 5447$ | 4.16 | \％ | 0.5 | BC212A | 0.1 | CA3006 CA3007 | 4.60 | CA3062 | 3.75 1.10 | LM341P12 | 0.80 0.80 | LM748－14 LM716 | 0.50 1.00 | LM7815CZ | 0.30 | SN76033N SN76110N | 2.35 1.30 |
| 2 N 1637 | 0.77 | 2－3638 | 4.17 | － N 4235 | 1.35 | 2N5448 | 0.15 | \％2U1 | 1．38 | BC212 2 A | 0.11 | CA3007 CA3008 | 4.15 2.55 | CA3084 CA3065 | 1.16 1.10 | LM341P15 | 0.80 | LM716 | 1.00 0.50 | LM78224C2 | 0.30 | SN76110N SN76115N | 1.30 1.65 |
| 2N1890 | 030 | 243639 | 0.48 | 2N4236 | 1.15 | 2N5449 | ${ }_{0}^{0.20}$ | CF201 | 1．70 | ${ }^{\mathrm{BC}} 213 \mathrm{Bb}$ | 0.15 | CA3012 | 1.65 | CA3068 | 3.80. | LM348N | 0.95 | LM911 | 0.50 | MC1035P | 1.50 | SNT6i16N | 1.80 |
| $\begin{aligned} & 2 N 1893 \\ & 2 N 1991 \end{aligned}$ | 110 | 7N3644 | 0 | ${ }^{\text {7 }} \mathrm{N} 42377$ | 1.55 | 2N5458 | 0.35 0.35 |  | 1.26 | ${ }_{\text {BC2 }} \mathrm{BC214} 4$ | 0.12 0.12 | CA3013 | 1.85 | CA3070 | 1.90 | LM358N | $0 \cdot 60$ | LM921 | 0.50 | MC1327P | 1.70 | SN76131N | 1.30 |
| 2N2193 | 0.50 | 7k3663 | 0.29 | SN4240 | 170 | 2N5555 | 0.65 | AF2TE | 0.8 | BC2141 | 0.11 | CA3014 | 2.20 | CA3071 | 1.90 | （M360N | 3.00 | LM923 | 0.50 | MC1330P | 1.10 | SN76226N | 1.68 |
| 2N2194 | 042 | 2k 3702 | 0.14 |  |  | 2N6109 | 0.55 | AF26．： | 0.05 | BC237B | 0.11 | CA3018 | 0.75 | CA3072 | 1.90 | LM370N | 3.30 | LM1303N | 1.15 | MC1352P | 1.20 3 | SN76227N | 1.30 |
| 2N2217 | 0.55 | 2N3703 | 0.14 | 2N4266 | 0.32 | 2N6122 | 0.44 | ASY28 | 1.30 | BC2388 | 0.13 | CA3018A | 1.10 | CA3075 | 1.70 | LM371H | $2 \cdot 35$ | LM 1304N | 1.52 | MC14336 | 3.65 | SN76228N | 1.55 |
| 2N2218 | 035 | 2h3704 | 0.14 | 2N4284 | 0.38 |  |  | ASY55 | 0.74 | BC239C | 0.11 | CA3020 | 2.20 |  |  | LM350K | 6.45 | LM1305N | 1.52 | MC1435G |  | SN76531N | 0.82 |
| 2N2219 | 038 | 273705 | 0.14 | 2N4286 | 4.32 | 2N6124 | 8.45 | W． 107 | 0.16 | BC256A | 0.2 | CA3020A | 2.50 | ca30 |  | LM373N | 3.35 | LM1307N | 1.22 | MC1439G | 1.75 | SN70532N | 1.55 |
| 2N2221 | 028 | ＋173706 | 0.14 | 2N4287 | 422 | 2N6125 | 8－4 | Sc108 | 0.16 | ${ }^{8 C 257 A}$ | 0.12 |  | 2.40 | CA3080A |  | LM374N | 3.35 | LM1310N | 2.10 | MC1440G | 1.85 | SN76533N | 1.30 |
| 2N2222 | 0.26 | ${ }^{2 N 3707}$ | 0.14 | 2N4288 |  | 2N6288 | 5．bs | 5109 | 0.15 | ${ }^{8 C 2588}$ | 0.24 | ${ }^{\text {CA33022 }}$ | 2.20 | CA3086 | 2． 0.50 | LM377N | 1.80 | LM1351N | 1.30 | MC1456G | 2.15 | SN76544N | 1.60 |
| 2N2270 | 0.49 | 2／3708 W3709 | 8． 12 | 2N4292 | ${ }^{2} 1$ | 2 S 702 | $3 \cdot 3$ | 201．3 |  | BC2598 | 0.17 0.85 | CA3023 | 2.20 | CA3088 | 1.87 | LM378N | 2.40 | LM1458N | 0.45 | MC1463R | 3.90 3 | SN76545N | 1.80 |
| 2N2388 | 0.27 | －1N1710 | ${ }^{1} 12$ | 2N4303 | 8.31 | 25703 | 3.93 |  | 0.22 4.22 | BC261A BC2628 | 0.25 | CA3026 | 0.7 | ${ }_{\text {CA }}^{\text {CA3088F }}$ | 1.87 2.90 | LM3799 | 4.25 | LM 4496 N | 197 | MC1438L | 3.35 3.16 | 8N76546N | \＄．59 |
| 2N2369 | 0.27 | －N3711 | 4.12 | 2 N 4342 | －1．6． | 40232 | 0.54 | 区1． | 4.21 | BC2628 | 0.25 | C41028A | 0.98 | CA30900 | 4.40 | LM380N8 | 0．8s | LM1800N | $1-94$ | MC1469R | 3.16 | SN76550－2 | 6．3 |
| 2N248 | $0 \cdot 30$ | SN3712 | 1.39 | 2 N 4401 | 4.28 | 40311 | 0.65 |  | F．22 | ${ }^{\text {BC2638 }}$ | 425 | LC10288 | 1.25 | CA3i30 | 1.06 | $1 \mathrm{M380N14}$ | 1．02 | LM1812N | 6－20 | MC1495L | 5.6 | SN76570N | 6．5） 1.85 |
| 2N2613 | 0．4 | 2N3714 | 4.65 | 2N4402 | － 26 | 40316 | I | C135 | 1.8 | BC26 | $4{ }^{\text {\％}}$ | － 01029 | 0.75 | CA3140 | 1.04 |  | $2 \cdot 10$ | LM182 | $1-16$ | MC1529G | 7.12 | SN76620AN | 1．85 |
| 2N2646 2N2848 | 1.15 | 2N37 | 170 | 2N4403 | 48 | 40389 | 7 | ค．13\％ | 1.91 | ${ }^{\text {BC3 }}$ C308B | 616 | C61029A | 0－90 | LOBST1 |  | LM381N | 1.69 | $1 \mathrm{ml}{ }^{\text {chen }}$ | 190 | MC4024P | 2.20 | SN／6650N | 1．26 |
| 2 ${ }^{\text {290 }}$ 294 | 0.7 t | 2N3794 | 0.21 | 2N4822 | 0.63 | 40408 | 0.12 | 3 ） | 17 | BC309C | 618 | C．6． 1030 | 1.50 | LM1 $14{ }^{\text {H }}$ | 2.75 |  | 1.12 |  |  | MM5314 | 4.62 | SN76660N | 0．65 |
| 2N2905 | 0.31 | 243819 | 0.36 | 2N4870L | 0.58 | 40440 | 0.74 | をじ30 | 0.44 | 8С327 | 42 | ［53030A | 2.20 | LM301AH | 0.50 | LM386N | 1.5 | LM184 | 190 | MM5316 | 4.62 | \＄N76666N | 0.87 |
| 2N2906 | 0.24 | 243820 | 0.39 | 2N48711 | 4.51 | 40512 | 1.70 | \％ 40 | 0.30 | BC328 | 020 | ${ }_{4} 41033$ | 3.78 | LM3018 | 0.30 | LM387N | 1．16 | LM1845N | 150 | MM5320 | 4.20 | 56100 | 2.75 |
| 2N2907 | 0.25 | 243821 | 0.95 | 2N4898 | 1.55 | 40594 | 0.87 | $3 \mathrm{c}, 141$ | 0.32 | BC337 | 020 | Cr， 1034 | 2.75 | LM304 | $2 \cdot 0$ | LM38N |  | LM1848N | 198 | NE555 | 6.33 | SL611C | 2.75 |
| 2N2923 | 0.17 | 743827 | 0.27 | 2N4901 | 1.65 | 40595 | 0.93 | ES．143 | 0.32 | BC414 | 0 r | C．23035 | 1.95 | LM307N | 0．010 | LM388N | 1.63 | LM1889N | 190 490 | NE556 | 6.65 | S． 612 C | 2.7 |
| 2N2924 | 0.17 | $\geq 43854 \mathrm{~A}$ | 0.30 | 2N4992 | 2.20 | 40673 | 0.80 | N： 141 | 0.13 | BC415 | ¢ 18 | CA3036 | 1.21 | LM308H | 1．89 | LM389N | 1.03 |  |  | NE560 | 4.51 | \＄1620C | 161 |
| 2N2925 | 0.19 | 2N3855 | 0.30 | 2N4903 | 2.75 | AC126 | 0.43 | E．149 | 0.15 | BC416 | 517 | CA3038 | 2.90 | LM308N | 0.45 | LM555cN | 0.33 | LM2907N－8 | 1.80 | NE561 | 4.50 | SL．621C | 1.1 |
| 2N3011 | 0.37 | 2N3856A | 0.19 | 2N4904 | 1.85 | AC127 | 0.48 | BC149 | 0.15 | BC547A | 0.13 | CA3038A | 4.10 | LM309KC | 1.95 | LM565CN | 1.39 | LM2987N－8 | 1.80 | NE562 | 4.50 | SL623C | 5.25 |
| 2N3020 | 0.75 | 2N3588A | 0.20 | ${ }_{2} \mathrm{~N} 4905$ | 2．49 | ${ }^{\text {ACLI28 }}$ | 0.43 | ${ }_{\text {B }} \mathrm{C} 153$ | 0.30 | ${ }_{8 C 5478}^{\text {BC5 }}$ | 013 |  | 0.77 | LM317K | 23.35 | LM7018 | 2.99 | LM3301N | 0.60 |  |  | SL640C | 4.4 |
| 2N3053 | 0.25 |  | 0.22 | ${ }^{2} \mathrm{~N} 4920$ | 6.83 0.30 | AC151 AC152 | 0.43 <br> 0.54 | P175．1 ACI5A | 0.30 0.15 | BC548 BC5498 | 4．93 | －A3040 | 3.76 | LM318N | 2.15 | LM701C LM 7026 | 2.99 0.81 | ${ }_{\text {LM3302N }}$ | 0.55 0.55 | NE566 | 1.35 1.90 | $\underset{\text { SL }}{\substack{\text { SL01C }}}$ | 4．4．4 |
| 2N3055 | 0.75 | ZH3866 | 1.98 | 2N5087 | ${ }_{6} 6$ | ACI53 | 0．69 | RC153B | 0.15 | ВС558 | ［． 13 | －${ }^{33041}$ | 1.65 1.65 | LM32005 |  | ， | 1.15 | LM3900N | 0.68 | NE558N | 1.95 | SL7016 |  |
| 2N3108 | 0.75 | 24x901 | 0.30 | 2N5088 | 436 | ACI 53K | 0.59 | RC15 $\mathrm{B}^{\text {B }}$ | 017 | BC559 | 0.15 |  |  |  |  | LM709 |  | LM3905N | f． 15 | NE571N | 4.85 | TAA233 | 1.33 |
| 2N3133 | 0.50 | 2M3904 | 0.18 | 2N5089 | 430 | AC176 | 0.54 | 日C16＊ | 038 | BCY54 | 48 | ${ }^{2} \mathrm{~A} 3045$ | 1.58 | LM320T24 | 2.15 | LM709 | 0.50 | LM3909N | 0.78 |  |  | rasiou |  |
| 2N3242 | 0.61 | $\underline{-143}$ | 0.18 |  |  | AC176K | 0.60 | 8－16．8 | 05 | BCY58 | －27 | －A3046 | 0.77 | LM3209 ${ }^{\text {P／}}$ | 1.15 | LM709．14 | 0.49 | LM391 ${ }^{\text {N }}$ | 1．10 | SAS570 | 0 | taas50a | 1．15 |
| 2N3250 | 0.3 | － | 8 | 2N5130 | 0.22 | AC187 | b． 59 |  | （0．13） | BCY70 | 121 | A3047 | 2.20 | LM320MP12 | 1．15 | LM710 | 0.67 | ［M4250CN | 1．30 | SAS580 | 2.40 | tas521 | ${ }_{1.15}$ |
| 2N3301 | 0.48 | 283962 | 0.95 | 2N5131 | 0.22 | AC187K | 4.65 | SC169B | 0.13 | BCY71 | $+25$ | CA3047A | 3.70 | LM320MP15 | 2．115 | LM710－14 | 0.64 | LM78L05CH | 0.85 | SAS590 | 2.40 | taA522 | 2.16 |
| 2N3302 | 0．39 | 2h4031 | 0.55 | 2N5137 | 0.22 | AC188 | － 54 | BC1708 | 0.19 | BCY72 | 4.13 | －A3048 | 2.45 | LM320MP24 | 1．15 | LM7IICN | $0 \cdot 12$ | LM78L12CH | 0.85 | SN76001N | 1.30 | taA550 | 0.41 |
| 2N3392 | 0.17 | 2N4032 | 0.65 | 2N5143 | 4.22 | AC188K | 5.65 | Q．1718 | 0.47 |  | 843 | ［A3049 | 1.93 | LM323K | 6.95 | LM723C | 4.75 | LM78L15CH | 0.85 | SN76003N | 2.38 | TAA560 | 2.19 |
| 2N3394 | 0.17 | 2\％1033 | 0.65 | 2N5180 | 4.58 | ${ }^{\text {ACF1 }} 1$ | 1.00 | ${ }^{8172 C}$ | 0.15 0.17 | ${ }^{80121} 8$ | 220 | A3050 | 2.66 | LM339N | 0.65 | LM723C－14 | 6.45 | LM78L24CH | 0.85 | SN76008KE | 1.60 | ta4570 | 2.25 |
| 2N3397 | 0.19 | ＜11036 | 0.72 | 2N5190 | 4.65 | ACY22 | 0. | － 1736 | 0.17 | B6131 | 055 |  |  |  |  |  |  |  |  |  |  |  |  |

[^5]
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A really first-class Hi-Fi Stereo Amplifier Kit. Uses 14 transistors including Silicon Transistors in the first five level with each channel resulting integral pre-amp with Bass, Treble and two Volume Controls. Suitable for use with Ceramic or Crystal cartriages. Very simple to modify to suit magnetic cartridge-instructions included. Output stage for any speakers from 8 to 15 drilled metalwork high, all parts supplied including dircuit board with high quamy ready drimed printed marked, smart brushed anodised aluminium front panel with matching knobs, wire solder, nuts, boltsno extras to buy. Simple step by step instructions enable any constructor to build an amplifier to be proud of. Brief specification: Power output: 14 watts $\pm 3 \mathrm{~m}$.
$\pm \mathrm{d}$
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ully detailed 7 page construction manual and parts AMPLIFIER KIT $\quad . \quad$. $\quad$ \&14.50 P. \& P. 80p POWER input components 33p ext $\mathbf{8 6 . 0 0}$ P. \& P. 95p CABINPT
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A solid state stereo amplifier chassis. with an output of atest high technology integrated circuit amplifiers with built in short term thermal overload protection. All components including rectifier smoothing capacitor, fuse. tone control, volume controls, 2 pin din speaker sockets \& 5 pin din tape rec. play socket are mounted on he printed Supplied brand approx. 9 w ${ }^{2}$, 1 max. depth. Supplied brand new \& tested, with knobs, the amplifier to be mounted horizontally or vertically) at only $£ 10-00$ plus 50 p P. \& P. Mains transformer with an output of 17 v a/c at $500 \mathrm{~m} / \mathrm{a}$ can be supplied at $£ 2.00+$ $40 \mathrm{D} P \& P$ if required. Full connection details supplied. All prices and specifications correct at time of press and subject to alteration without notice.
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## Easy

Verospeed have announced the introduction of two "Zero Insertion Force" (AB 052) di.i.p. sockets, available in $16-$ and 24-way configurations. They are designed to eliminate mechanical damage to expensive integrated circuits during test, burn-in or programming operations. The i.c.s may be inserted or removed with zero force on the leads.

When the lever is operated the socket exerts uniform force on the leads and compensates for varying lead thicknesses. They are priced at $£ 5.36$ and $£ 7.10$ respectively and are available from: Verospeed, Barton Park Industrial Estate, Eastleigh, Hants SO5 5RR. Tel: (O703) 618525.


## Wire-wrapping

OK are probably the only company who produce wire-wrapping equipment specifically for the hobby market and their latest kit HW-KI brings power wire-wrapping within economic reach of the home electronics enthusiast.

The main item in the kit is a newly designed battery-operated wirewrapping gun, based on the design of the company's industrial units. It is for use with $0.6 \times 0.6 \mathrm{~mm}$ mini-wrap terminals and has a bit and sleeve to give modified wrap. The tool is also self indexing and has a back force device to prevent overwrapping.
It is powered by two NiCad batteries, the two batteries provided having a year's guarantee, and a mainsoperated charger is included. Also in the kit is a handy 'pocket-sized' wire dispenser, containing 50 ft of 0.25 mm

## Mini-drill

Recently introduced by Boss Industrial Mountings Ltd. is a mains drill plus accessory kit housed in a specially designed carrying/presentation case with transparent lid.

This small but powerful 220/240V a.c. BIMDRILL is supplied with 4 collets capable of accepting tools with shanks of $1 \mathrm{~mm}, 2 \mathrm{~mm}, 2.4 \mathrm{~mm}$ and 0.125 in diameter, and will readily drill brass, steel, aluminium and p.c.b.s.

Complete with spring-loaded on/off switch and a 2 metre long cable fitted with a 2-pin DIN plug, this BIMDRILL has a fully insulated, high impact, ABS body, weighs less than 250 g and has an off load speed of approximately 7500 r.p.m.

The accessories include 20 assorted twist drills, mops, burrs, grinding wheels and mounted points, all individually and securely located within the $230 \times 130 \times 50 \mathrm{~mm}$ case. Priced at $£ 22.14$ which includes VAT and P\&P, the kit is available from: Boss Industrial Mouldings Ltd., Higgs Industrial Estate, 2 Herne Hill Road, London SE24 OAU. Tel: 01-737 2383.
wire. The other item is a hand tool which can strip wire, wrap and unwrap.

Suggested retail price of the HW-KI is $£ 46.28$ (inc. VAT and carriage). $O K$ Machine \& Tool (UK) Ltd., 48a The Avenue, Southampton, Hants SO1 2SY. Tel: (0703) 38966/7.


## Drill-Mills

A circuit board without a track? A single Stahler drill-mill bit simultaneously drills holes and mills isolated pads in plain copper clad board to produce circuit boards without etching.

Drill-mills are designed to perform two separate functions-Isolated-pads drill-mill drills a hole and mills a circular moat leaving an electrically isolated copper pad. Insulated-spot drill-mill also drills a hole, but instead of milling a moat, it milis out a disc
from the copper cladding around the hole. The application is to provide clearance spots for ground-plane mounting of d.i.p. sockets, feedthrough sockets, etc.

In each case the tools comprise a replaceable high speed or carbide twist drill, either $0 \cdot 0400$ in diameter or 0.0292 in diameter. Each tool is also available in three pad milling sizes, 0.01 in, 0.15 in and 0.20 in . Prices and a free catalogue are available from: Carel Components, 40-44 The Broadway, London SW19 1SQ. Tel: 01-5407186.


by Eric Dowdeswell G4AR

An interesting letter from Les Light G3KDL of Wembley, Middx, prompts me to write a few words for the reader who may be reading this column for the first time, perhaps after having had interest aroused elsewhere by contact with amateur radio.

Les received a letter from a young but obviously enthusiastic listener simply saying he had heard Les on the 80 m band. Time and frequency were given, but little else, except some details of the receiver and aerial. Now, with the best will in the world, the report was not of much use at all, but Les, no doubt thinking back to his own early days in amateur radio, did send a QSL.

It is naturally very exciting for the newcomer to listen to amateurs for the first time especially when they realise the comparatively low power of the amateur, possibly being accustomed to the high power stuff on the BC bands. So off goes a report such as quoted above, with a request for a QSL, and probably no s.a.e!

With the advent of so much commercial gear and highgain beam aerials, worldwide working by amateurs is commonplace and the excitement of working and then getting a QSL from some DX spot has waned somewhat. QSLing had decreased over the years to the extent that some amateurs don't even bother to have cards printed!

So the report from a listener, even if from the other side of the world, is of little use, generally speaking. It has got to be something unusual in the reception, or perhaps the other fellow is using very low power, before it is worth sending a report. I would counsel any beginner to first get the Guide to Amateur Radio from the RSGB to see what it is all about and to study, in particular, the section dealing with QSL matters.

The cost of QSL cards and postage today is enough to deter anyone from the hobby, so if you must send a report and request for a QSL, then make it easy for the other station by sending a s.a.e., or International Reply Coupons for overseas stations, and a report worthy of attention. Don't use these silly $5 \times 3$ in envelopes! Send something a bit bigger. Having sent tens of thousands of QSLs out when I was ST2AR I rate the small envelope as the biggest annoyance of all!

Collecting QSLs can be fun, or used to be, but if there is any intention of going on to get a transmitting licence I would say the best policy is to spend the money on the receiver and associated gear rather than on QSLs and QSLing. You'll need every penny for the entrance fee for the RAE and the ticket itself. Then you can have your own QSL card, and receive them from stations you work, when they really mean something.

## General Notes

An interesting note from Owen Frame G4EIF of Reading, Berks, on the mysterious H44LW mentioned in the December issue. Owen confirms that it is the Solomon Islands, as he has a QSL for an s.s.b. QSO. QTH is Box 19, Honiara. Owen added, that he used to be ST2WF in Khartoum in 1934 with the RAF, with 80W of c.w. from 12 V accumulators! As he says: "happy days".
N. Eddy writes from Truro, Cornwall, for the first time although he has been reading $P W$ for many years. He is a member of the Cornish ARC and sports an AR88 and BC 348 with main interest on 14 MHz . He also has a 62 H receiver which he'd like to convert for 2 m operation, if any reader can help him on mods. Write to: Little Tregadles, Laity Moor, Ponsanooth, Truro, Cornwall, all expenses paid of course. John Bell BRS40279 in Melksham, Wilts, had "quite an amazing response" to his request for info on the AR88, particularly from Michael Swain G8MMP. John recently got hold of an Eddystone EC10 and, as an instructor to mentally handicapped people, he is going to try to teach elementary geography using the EC 10 to demonstrate the reception of BC stations from various parts of the world. Apparently a fellow instructor has had some success this way.

Newcomer to the column Derek Brabrook of Laurgharne, Dyfed, Wales, has been using a domestic type Pye set for a couple of years or so, as it goes up to 26 MHz . However, he finds his 223 ft aerial "difficult and poor''. I'm not surprised! It must be overloading the front end I imagine on the stronger signals, as I doubt whether there is any r.f. gain control, and selectivity can't be much good on such an old set. Instead of blaming the set, I suggest keeping the aerial and getting a more modern set!

In Oswestry, Salop, David Wyatt aged 14, has acquired a BC348 and promptly found KC4USX in Antarctica on 20 m s.s.b. His main regret is that it does not go above 18 MHz so he can't listen to the 10 and 15 m bands. Again, a case for a converter. David offers to reply to any readers who'd like to write to him on the BC348 at 11 Prince Charles Road, Oswestry. You might be overwhelmed, as the receiver was a very popular one OM! David is also learning the code from records and threatens to send in c.w. reports before long. We are certainly short of them of late; hint, hint, to other readers!

## Round the Bands

Very little in the way of logs this month, unless the season's rush on the PO has held them up in the pipeline. Could be everyone's waiting to see if Father Christmas has remembered them with a nice, shiny, new receiver! Bernard Hughes BRS25901, who hasn't written in for a while from Worcester, liked my article on receiver accessories and wants more! Drop a line to the Editor OM! In spite of a new 20 m dipole Bernard found the most interesting DX on other bands with his Drake receiver. Latest QSL received of any note was from PYORO on St Peter \& Paul Rocks. Worthy of note in his log were HS1ABD, ST0RK (!) VR3AH and XT2AT on the 10 m band, KJ6BZ VP2DAY on 15 m and HR1HMV, YS1RRD and 4 W 1 BC on 40 m , all s.s.b.

From Leigh-on-Sea, Essex, Ian Marquis A9140 keeps up the good work keeping his ears open on all the h.f. bands from Top to Ten with stuff like JA2EMU TR8BA on $80 \mathrm{~m}, \mathrm{KL} 7$ IRT and ZL4AV on 40 m , and FP8, FR7, PZ5, TU2, ZD8 and 7P8 on the 15 m band, again all s.s.b.

A letter from Bernic Crockford ZS1BW mentions a contest for amateurs and SWLs to celebrate the 150th anniversary of the University of Cape Town, from Saturday Feb 17 to Sunday March 4, 1979, with operation from 0600 to 2000 on Sundays, and 0700 to 1000 and 1500 to 2000 weekdays. Likely frequencies are $7050,14210,21200,28580$, subject to QRM. Contact/log ZS1UCT plus two other ZS1s for award. Details from: Awards Manager ZS1MO, PO Box 5100, Cape Town 8000 , Rep. of South Africa.

## Club Activity

Much more from the clubs this month than individual readers! RAIBC secretary, Frances Wooley G3LWY, reports that the Strumech tower raffled at the Leicester Show was won by disabled member Shirley Hesketh G4HES! Shirley has already coached to success several blind girls at Chorleywood College, for the RAE. Club net 3750 kHz s.s.b. 1000 Tuesdays and 1400 Wednesdays. The Cheshire Homes net is on 3650 to 3700 kHz on Thursdays 1330.

The Silverthorn RC meets Fridays 1930 at Friday Hill House, Simmons Lane, Chingford, London E4, with details from: C. J. Hoare, at that address. Newsletter "Spurious" would be considerably improved with less large cartoons and virtually useless photographs, and more details of events to come! Surely you must have a winter programme chaps? Tars Talk, journal of the Torbay ARS reveals that G3LHJ will give a slide show on Feb 24, with the society's annual dinner being held on March 10. Meetings are held at Bath Lane, Torquay (rear of 94 Belgrave Road). Details from: F. Bolton G3VTQ, 2 Lower Coombe Road, Blindwell Park, Kingsteignton, Newton Abbot, Devon.

Stevenage and District ARS continues to meet at British Aerospace, Gunnels Wood Road, Stevenage, Herts on first and third Thursdays at 2015. March 1 sees talk by GB3HR repeater group on proposed 23 cm and 10 GHz beacon, while the 10 th sees a visit to the VHF Convention at the Winning Post, Twickenham, Middx. AGM is on the 15 th. The West Kent ARS report in for the first time with details of meetings held at 2000 at the Adult Education Centre, Tunbridge Wells, on Fridays, with a junk sale on March 2. On March 30 an interesting talk by Tony Tory on microprocessors in amateur radio. Informal chit-chat and code practice on Tuesdays at the Drill Hall, Victoria Road, says Brian Castle G4DYF, 6 Pinewood Avenue, Sevenoaks, Kent, who will supply details.

A new QTH for the Swansea ARS at the Sketty Park Sports and Social Club, Aneurin Way, Sketty Park, Swansea, on alternate Tuesdays, Feb 6 and 20 et seq. New members welcomed with open arms and a pint! Ring Peter Jones GW4GRI on Swansea 873986 for info or write to: 27 Gorwydd Road, Gowerton.
J. Bazley G3HCP, President of the RSGB, will be guest of honour at the 31st annual dinner of the Sutton \& Cheam RS, taking place at the Woodstock Hotel on Saturday March 24. Details and tickets from: G. W. Brind G4CMU, 26 Grange Meadow, Banstead, Surrey.

Events to be held by the Cheltenham AR Assoc include: a constructors contest on Friday Feb 16 and a talk on i.c.s by Eric Hibbett G8LAY on Thursday March 1, with G4BSO and G3SSO discoursing on aerial planning permission on the 16th. Meetings at the Old Bakery, Chester Walk, Cheltenham at 2000. Info from: G. Martin G3IER, 88 Tennyson Road, Cheltenham, Glos.

## Log Extracts (All s.s.b.)

B. Hughes:-10m HS1ABD KZ0DX P28NKV S79MC ST0RK VS6FI VR3AH XT2AT 15m D68AD KJ6BZ TJ2AP VP2DAY 40m HR1HMV HK5BCI YS1RRD ZP5LX 4W1BC
I. Marquis:-10m HP1PJ VP2MBD ZL3AAX 15m CT2BB FP8DX FR7ZN PZ5AA TU2FH ZD8RG ZF2AG 7P8BC 20m VP2DAO ZE3JO 40m HK5DUS KL7IRT ZL4AV 80m FP8DX JA2EMU LX1ST TR8BA 9H1EU

Remember, all logs and letters by 15 th of the month.



## MEDIUM WAVE DX

## by Charles Molloy G8BUS

It may be arguable, whether humility is the greatest virtue that a m.w. DXer should possess, but there is no doubt at all that selectivity is the attribute to look for in a receiver to be used for serious medium wave DXing. It is not that other factors such as sensitivity, stability, scale accuracy, freedom from overloading and cross-modulation, are unimportant. The present overcrowded state of the band, means that you should have the facility to winkle out DX that is close to a strong local station, and this means using a receiver that has good selectivity.

What is selectivity? It is the ability a receiver has to separate stations that are close to one another in the band. You can regard a receiver as a window into the frequency spectrum, the width of this window depending on the degree of selectivity. A narrow window means narrow (good) selectivity. It would be an easy matter, if all you had to do was to separate adjacent carriers, since a receiver with high selectivity would do the trick. Unfortunately the programme (modulation) requires space, the amount of space depending on the highest audio frequency. If it is 3 kHz then a 6 kHz bandwidth is required, i.e., 3 kHz above and 3 kHz below the carrier with the double sideband system currently in use. If bandwidth is reduced in order to reduce QRM then the audio range is reduced too, and in an extreme case speech becomes unintelligible.

## How Selectivity is Measured

The handbook of a well-known receiver quotes the selectivity as 4 kHz at -6 dB and 18 kHz at -40 dB . What do these parameters mean? The first is the important one. 6 dB is shorthand for 6 decibels and -6 dB means "one half of the original value", so the signal at the sides of the 4 kHz window is half that at the centre. Selectivity is invariably measured at the 6 dB points. The statement 18 kHz at -40 dB is less important. It means simply that at 9 kHz on either side of the carrier the signal will be 40 dB down which is $1 / 100$ th of its original value. Ideally the signal would be zero at the sides of the window, but in practice it falls off gradually.

## What to Look For

If you want to judge a receiver's ability to separate stations then look for the bandwidth at the 6 dB points. It will be found under "Selectivity" in the handbook or specification. Sometimes it isn't mentioned at all so one can draw the appropriate conclusion. Occasionally on imported receivers the bandwidth is given as a plus or minus figure such as $\pm 2 \mathrm{kHz}$. Multiply by 2 , as the real bandwidth is 4 kHz ! I recently studied a one-page "spec" for a receiver currently available and after some searching found Selectivity near the end. It was 4 kHz at the 6 dB points which is not very good for DXing. A few years ago I tried an experiment with CJON (now CJYQ) on 930 kHz using a bandwidth of 2.4 kHz . It could be heard easily between the Europeans on 926 and 935 . When the bandwidth was in-
creased to 5 kHz , CJON just disappeared, as the stations on either side spread out to meet each other.

My BRT400 communications receiver has a sixposition selectivity switch which gives bandwidths of 0.5 , $1.0,2.0,5.5,9.0$ and 13.0 kHz at the 6 dB points. Normally I use 2 kHz when tuning around and this is increased to 5.5 kHz or even 9 kHz , QRM permitting, if I want to listen to the programme. A receiver with fixed selectivity must compromise between the needs of the DXer and the listener, and I would question whether such a receiver should be entitled to use the term "communications". A personal view that many will disagree with. It is a pity that the Q Multiplier, referred to recently by Eric Dowdeswell, has gone out of fashion as this simple device provides an easy means of obtaining variable selectivity, and it was incorporated in a number of moderately-priced receivers a few years ago.

If you do have a receiver with narrow selectivity then it is possible to use this facility and still hear the modulation. If the programme quality deteriorates as selectivity is increased, then detune slightly, away from the offending QRM. Speech quality will immediately improve, as your window will now look out on only one of the two sidebands. It may come as a surprise to some DXers to find out that the programme is actually carried twice on an a.m. double-sideband system. There is a sideband complete with programme on each side of the carrier and it is only necessary to tune to one of these to extract the modulation. You could double the number of channels on the medium waves by suppressing one sideband but there are problems in doing this.

## The New Band Plan

I remained up late on the night of the big change over last November, just in case I might be missing something, and was well rewarded with the feeling that I was listening to history being made. After hearing the short announcement on 200 kHz , at midnight, I tuned round the medium waves to be greeted by tuning notes on nearly every channel. The change at the h.f. end of the band was striking. Where the German power-house on 1602 had been, only a few minutes earlier, was now the third international common frequency with nothing to be heard but a burble. That evening I hunted around for any DX between channels, but all I could find was an unidentified Arab on 1570 . It looks as if Asiatic DX has disappeared, in the evenings at any rate.

## Readers' Letters

Steve Whitt (Cambridge) is interested in QSLs and he reports that he has had $100 \%$ returns from the 24 US stations he has reported to so far, but from further south, Radio Margarita on 1020 and Radio Coro on 1210 have not replied. This is a common experience as Latin Americans are notoriously difficult to QSL. One approach is to write a personalised letter to the station giving details of one's self, and perhaps a photo of the shack, or the locality. WEVD 1330 broadcasts in Hebrew, and Steve wonders how he can compile a report, as the programme material is meaningless to him. Make a tape of the DX and if you cannot get it translated then send the tape to the station. Many DXers send tapes instead of reports though it is rather an expensive way of doing it.

Two other questions come from Steve: 1. What is the station on 593 (pre-Geneva) underneath the West German that relays the BBC World Service? It is probably crossmodulation which can occur in the ionosphere as well as inside the receiver. If you listen on the open carrier of any
high power European, just before sign-off, you may hear this effect. There are so many megawatts floating around the ionosphere these days, that all sorts of weird effects can be observed. 2. What is the easiest English speaking DX from the Caribbean? None is easy. Try ZDK Antigue on 1100 or Radio Paradise, St Kitts on 1265, but you will have to stay up late to hear them.

## DX Heard

John Faulkner (Mansfield) reports hearing 19 Canadian and 13 US stations with his Trio 9R59D receiver and 40 inch $P W$ loop. Toronto was logged on three points on the dial with CBL on 740, CJBC 860 and CFRB on 1010. Others include WHAM Rochester, NY on 1180 and Fort Wayne, Indiana on 1190 kHz .

An HMV domestic receiver and loop are in use in Birmingham by John Dennis Court and his log contains two not-so-often heard stations in Newfoundland; VOWR on 800 at 0030 and VOAR on 1230 at 0200 . There are still a few outlets in Newfoundland that use the old prefix " V " in the call sign, VOCM on 590 being the one that is usually heard. Noel Cosgrave (Dublin) has a Mullard MAS1659 receiver and 36 inch loop, which brought him Radio Belgrano, Buenos Aires on 950 kHz at 0255 , Radio Tupi in Rio de Janeiro on 1280 at 0315 and CB57 Santiago, Chile on 570 kHz .


## SHORT-WAVE BROADCASTS

## by Charles Molloy G8BUS

Skip is a term frequently used by DXers and it is one that may confuse the newcomer to the hobby. Very briefly it refers to a zone around a transmitter where little or no signal can be heard. The ground wave travels only a few miles from a short-wave station before petering out, while the nearest place where the sky wave returns to earth may be hundreds of miles away. The distance between the two is the skip distance. Anyone situated nearer to the transmitter than the point where the sky wave returns, will be in the skip zone and consequently will not hear the station at all, unless of course he is within range of the ground wave. This is the reason why BBC short-wave transmitters in this country are not heard within the UK. The sites of these transmitters incidentally are at Daventry in Northamptonshire, Skelton in Cumberland, Wooferton near Ludlow and Rampisham in Dorset.

## The Sky Wave

Great stuff, you may say, but why is it that the sky wave fails to come back to the earth inside the skip zone. To understand this you have to look at what happens to a wave sent up vertically. If the frequency is low enough then the signal will travel some distance into the ionosphere before being returned to the transmitter. A higher frequency will go up a bit further before coming back, and if frequency is increased gradually, a time will
come when the wave will travel right through the ionosphere and off into space. The highest frequency to be returned is called the critical frequency and is usually designated $\mathrm{f}_{\mathrm{c}}$.

Vertical radiation is not much use for broadcasting so in practice a lower angle is employed. A wave travelling at an oblique angle will have to travel a greater distance through the ionosphere than a vertical one, before coming out at the top, so frequencies higher than the critical frequency can be used at low angles. The highest frequency that will be returned will be when the wave is at a low angle, just above the horizon, which is the case for long distance transmission and this frequency may be three or four times $\mathrm{f}_{\mathrm{c}}$. The Maximum Usable Frequency (m.u.f.) for any particular angle can be calculated from $f_{c}$ which in turn is fairly easy to measure. One final point. To get the maximum signal into the target area it is desirable to transmit as near as possible to the m.u.f., and any radiation at a higher angle than required for this distance will therefore penetrate the ionosphere and be lost. This is what happens to the signal that is missing in the skip zone.

## Propagation and DXing

About 4000 km is the maximum distance that can be covered by a single hop from low angle radiation and in this case the skip will be great. If the target is nearer than 4000 km then a higher radiation angle will have to be used at the transmitter and a lower frequency will be needed as well. Obviously the skip will also be lower. Reception areas further away than 4000 km will be reached by multiple hops, the wavefront being reflected from the earth's surface back into the ionosphere after the first hop. So, at any particular time of day, season of the year or period of the sunspot cycle, all of which affect the critical frequency, the highest frequencies available will be used for long distance working and lower ones for short distances. There is no use looking for your favourite European local on 13 metres, for even if the band is open the skip will be too great.

## 19 Metre Band ( 15 MHz )

Following on from last month we will now have a look at 19 metres whose limits are 15100 to 15450 kHz , though there is some spread on either side. 19 metres is mainly a daytime band with world-wide reception being possible. Look for Vietnam on 15012 kHz , Teheran 15 084, Japan 15 105, RSA 15 220, New Zealand 15 130, Tanzania 15435 . Some medium range DX can be heard during the day, such as Norway on 15 175, Morocco 15 195, Sweden 15240 and Finland on 15 265, but these disappear as dark approaches and the m.u.f. falls.

During the evening Latin American DX can be heard and although at first sight this may appear surprising, it should be noted that the signal path is from the south-west from the southern hemisphere, where the greater part of the route will still be in daylight. DX heard regularly in the UK includes: Chile on 15115 , Brazil on 15 145, Chile on 15 150, Radio el Mundo Argentina on 15 290, Radio Nacional Colombia on 15335 , R. Mexico on 15385 , Venezuela on 15400 , and from the same area, Radio Grenada on 15105 kHz .

## Readers' Letters

The MCR1 wartime receiver is mentioned again by Trevor Goodenough, who has two of them which he uses regularly with a 300 ft long wire and an a.t.u. DX heard with this set-up includes Radio Japan and Radio

Reports on the various bands are welcome and shouid be sent direct, by the 15 th of the month, to:
AMATEUR BANDS Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey KT21 2TW. Logs by bands, each in alphabetical order.
MEDIUM and SW BANDS Charles Molloy G8BUS. 132 Segars Lane, Southport PR8 3JG. Reports for both bands must be kept separate.
VHF BANDS Ron Ham BRS 15744 , Faraday, Greyfriars, Storrington, Sussex RH2O 4HE.

Australia. Trevor is anxious to get hold of a copy of the circuit diagram or the manual for the MCR1 and if anyone can help, would they please contact Trevor direct at 8 Glencraig Terrace, Fenwick, Ayrshire. Postage will be refunded.

The logging of Radio New Zealand on 15130 kHz at 0630 by S.I. Fass in the December issue, is referred to by T. W. G. Elsenham, who points out that the RNZ schedule says that this channel closes down at 0450 . Schedules are always changing and cannot be relied on. Mr Elsenham goes on to refer to the pre-war BBC weekly World Radio. This had a feature covering stations heard by readers which included details of the programmes, and he wonders if we could do the same here. Sounds a very good idea. There are some interesting programmes to be heard on the short waves, usually from the less conspicuous broadcasts. Details of programmes from DX stations should be brief and give the date, time, and frequency (if known).

## DX Reported

An unusual $\log$ of 60 metre DX, all of Venezuelan stations, has been received from Leon R. Sin Sun of Aberdeen, who used a Sony CF9JOS between 2300 and 0400 to pull in Radio Lara on 4800 kHz , Bolivar on 4770 , Universo 4870, Juventud 4900, Yaracuy 4940, Rumbos 4970, Ecos del Torbes 4980, Barquisimeto 4990, Continente 5030. Programmes heard included local popular music, folklore and interviews which would be of interest to DXers who understand Spanish. Leon does not use any special aerial as he is rather short of space. Try the outside TV aerial if you have one or alternatively a Joystick plus a.t.u. might be useful, but consult the manufacturer, Partridge Electronics, who advertise in $P W$, to make sure that a Joystick can be used with your Sony receiver.

An FRG-7 plus 60 ft loft aerial and a.t.u. are in use in Wigan by Jim Edwards, who reports hearing KGEI, La voz de Amistad, on 9575 at 1030, Baghdad on 9745 at 2130, Lagos Nigeria on 11770 at 0630, WINB Red Lion on 15185 at 2000. Jim reports that he is now a monitor for Family Radio WYFR, which he finds an interesting diversion when he is not $D$ Xing.

## Australian Domestic Short-wave Stations

Lex Arnold, Hemel Hempstead (R107 ex-WD communications receiver and 9 ft indoor aerial), sends some notes about the domestic service of the Australian Broadcasting Commission. He reports that ABC Melbourne has been heard regularly between 1000 and 1300 on 9680 kHz . From past experience Lex suggests that DXers should try for stations in this network between late November and mid-April, especially during February and March. Listen for VLH9 on 9690 between 0730 and 1300, VLH11 on

11880 from 1900 to 2115, VLH15 on 15230 from 2230 to 0715 , all from Melbourne; VLM4 on 4920 and VLQ 9660 between 2200 and 1400 , which are in Brisbane; VLW6 on 6140 and VLW9 on 9610 from 2200 to 0100 and 1000 to 1600 , both in Perth. Readers who think that Radio Australia is not really DX should try for some of these domestic stations. There is also Port Moresby in New Guinea with P2T9 on 4890 between 0715 and 1400 and again from 2000 to 2200 plus P2T9 on 9520 between 2215 and 0700.


## by Ron Ham BRS15744

First, our sincere congratulations to our reader Robin Bellerby G3ZYE, Hove, Sussex, on his election to the council of the Radio Society of Great Britain. Robin, a member of the Brighton and District Radio Society, the Mid-Sussex Amateur Radio Society and the Sussex Repeater Group, is actively interested in all bands from 160 m to 70 cm . News of his election victory came just after the Brighton and District RS had made Robin the first recipient of their trophy, the Bill Pitfield Memorial Award, given for meritorious service to amateur radio by a club member. Like all RSGB council members, Robin has some hard work ahead, what with society affairs, conferences, conventions and WARC 79, the outcome of which may well affect all of us. Congratulations also to $\mathbf{D}$. J. Stewart, ex-G8MZP, on passing his Morse test and now sporting the call sign G4HSY.

## Solar Activity

Nigel Fisher, Goff Gill, Robin Knight, Peter Mynheer and Chris Peeder have recently completed a 60 MHz radio telescope for the radio section of the South-East Essex Astronomical Society, and were delighted when their new instrument, built with the r.f. and i.f. sections of an old TV receiver, recorded solar activity on December 10, 11 and 12 , as did the 136 MHz telescope of Cmdr Henry Hatfield, Sevenoaks, and the 146 MHz receiver which I use. Henry recorded a slight increase in solar radio noise during the afternoon of November 24, and again on the 25 th and, using his spectrohelioscope, he observed a large prominence on the north-west limb of the sun which lived for about 24 hours and he saw another on the east limb on the 26th.

It must have been this solar activity which caused the blackout on 80 m , reported by Alan Baker G4GNX, Newhaven, at midday on the 25 th, the aurora during the same afternoon and the ionospheric disturbance reported by the BBC World Service at 0215 on the 26th. At 1015 on the 25 th, I did not hear any International Beacon Project signals on 10 m and Alan said that on 80 m , even the strong shipping stations which share the band were weak and, from Applecross, Western Australia, Anthony Mann writes: "November 26 , absolutely dead, nil above 28 MHz all day".

## Aurora

The land lines were soon buzzing with an alert once Roy Bannister G4GPX, Lancing, John Cooper G8NGO, Cowfold, and Dermot Cronin G4GRO, Royal Sovereign, heard auroral signals on 2 m . Dermot worked GI and GM and G4GNX, who was soon in on the action, heard PA3. Dave Cox G80PR, Andover, Hants, worked stations in 5 different QRA locator squares during the event which included 3 GMs and a GI. For John Branegan GM8OXQ, Saline, Fife, this aurora was a novel event because, at 1135, he had a QSO with SM4IVE and soon after with G8OGD via OSCAR 8J. The SM thought an aurora was starting over central Sweden and the G8 told John that his signal had an auroral tone.

At 1320, John again contacted SM4IVE via OSCAR 8J, who reported that the auroral activity in Sweden and Finland was fading and thought to be heading towards Scotland. Between 1340 and 1400 John was receiving auroral signals from the Russian satellite, RS-1, and when he left the satellite he found a full scale aurora affecting the 2 m band, which he monitored until 1840. During the period he heard tone "A" signals from the 2 m beacons in Cornwall GB3CTC, France FX0THF, Germany DLOPR, Lerwick GB3LER, Northern Ireland GB3GI, and Wrotham GB3VHF, in addition to amateur stations from EI, G, GD, GI, GM and SM. GM8OXQ said there was another, but much weaker, aurora on the 26th but only the northern GMs beyond Aberdeen could use it. Anthony Mann says: "It would be lovely to witness an aurora here", and thinks that the only two visible there were during the $1946 / 7$ sunspot maxima. It really is a wonderful sight Anthony, I will never forget the beauty of the one I saw from southern England, following a big solar event in August 1972.

## The 10 Metre Band

M. Mrzyglod is delighted with the performance of his new FRG-7 and with his 10 m ground-plane aerial, he has been logging the IBP signals from Bermuda VP9BA, Canada VE3TEN, Germany DLOIGI and s.s.b. signals from amateurs on both sides of the Atlantic. M. Mrzyglod is looking for a circuit which he thinks was published in Practical Wireless some years ago for a Band I converter. If anyone can help, please let me know. Neil Clarke BRS 34306, Nottingley, York has been enjoying the DX on 10 m and logging A4XFA, OX3CO, and VS6FI.

Like myself, John Branegan has found that the signal from the Bahrain beacon A9XC, was very consistent between November 19 and December 15, in fact, I heard it every day except during the blackout on November 25. John, Alan Baker and I, also heard signals from 5B4CY, DL0IGI, N4RD, VP9BA and 3B8MS. On his newly acquired Eddystone 770R communications receiver, John can tune through Bands I and II, and has already observed the effect of F2 conditions on the American and European stations which operate between 30 and 40 MHz . He is looking forward to using it during the coming sporadic-E season.

## From Down Under

Anthony Mann reports strong sporadic-E disturbances on November 16 and 28, producing Band I colour on an indoor aerial and m.u.f. lapping Band II. On November 18: "The best F2 opening so far", says Anthony who received strong signals from the BBC , Channel B 1 sound, 41.5 MHz , French Channel F2 sound, 41.25 MHz , a watchable picture from a South Korean station, strong Chinese video, 57.770 MHz , strong Russian TV sound,
east and west Malaysian video, and strong signals from Japanese amateurs in the 6 m band. We are all very envious Anthony.

## Tropospheric

During the morning of November 22, Alan Baker worked F1EZP and Mick Senior G4EFO, Horsham worked F6DOP, both through the Brighton repeater GB3SR, R3, and for most of the 22nd and 23rd I heard GW stations working through GB3BC, R6. At 2201 on the 19th, Alan had an unusual contact which lasted 28 minutes, through GB3SR; he worked N2AFE/MM, Ralph, from 5th Avenue, Brooklyn, New York, who was in the engine room of the Export Patriot in the English Channel, and as soon as other people realised what was on, suddenly Ralph had 19 stations after him.

A more extensive v.h.f. opening took place on December 5, 6 and 7, during which time I received strong signals from both the Bristol Channel and Dover 2 m repeaters. Throughout the 6th I received good pictures from the IBA transmitter at Lichfield, Channel 8, 189 MHz , and signals from several continental broadcast stations in Band II. At 0030 on the 6th, John Cooper phoned G4GNX to say that he had worked a station through a Belgian repeater, and G4GNX reported hearing ONOOV on R4. During the morning, GU2FZC, St Peter Port and F1EBE, Rouen, worked through the Brighton repeater and in the afternoon GB3BC was putting a strong signal into east Sussex. Periodically, throughout the day, Band V TV suffered from co-channel interference and in the evening, G6GL, Radlett, Hertfordshire, had an effortless QSO with G3TIR in Crawley, Sussex, via the Hampshire repeater, GB3SN, R5. Signals from the French repeater on R9 were heard on December 8 by G4GNX and G4GPX, and G3TRY, High Wycombe worked a station in Yorkshire through extreme QSB.

## Satellites

On December 1, our satellite expert, John Branegan, wrote: "Being a thorough optimist I still go on OSCAR 8A despite the high m.u.f., and though the ionosphere must be as thick as treacle, I have twice got across to W2BXA in New Jersey". The W2 told John that he was the only European he had heard for some time. "On Mode J at weekends it is a very different story", says John, who in 10 QSOs in November, worked 4 stations, W9KDR Conn, WB2OXJ New Jersey, WA3ZHW Penn, and VE2LI Montreal. By the end of November, John, by working his first east-German, DM2DIN, s.s.b., made his score 27 countries on 8 J and pushed his total up to 35 countries via satellite.

## Contests

As usual the RSGB have catered for the v.h.f./u.h.f. enthusiasts in their contests calendar for 1979, and reports will be welcome from any of our readers who take part in either the transmitting or receiving sections of any of the following events:

| March 3 and 4 | 144/432MHz and SWL |
| :---: | :---: |
| April 8 | . 432 MHz open and SWL |
| April 29 | . 70 MHz Open |
| May 26 and 27 | . 144 MHz Portable |
| July 7 and 8 | VHF NFD |
| August 18 and 19 | . 70 MHz Open |
| September 1 and 2 | 144 MHz Open and SW |
| October 21 | 70MHz Fixed |
| December 2 | 144 MHz Fixed |



by RON HAM

"'THE TWO ERNS"



One of the fascinations for me when writing my column is to realise that my readers are using wavelengths from 10 metres through to 3 centimetres, which in frequency terms, is 30 million to 10 thousand-million hertz. It is men like Ern Downer G8GKV of Worthing (right), and Ern Hoare, G8BDJ of nearby Southwick (left), known locally as "The Two Erns", who are pioneering the microwave end of the radio frequency spectrum.

Ern Downer's schoolboy interest in amateur radio was fostered by a neighbour, the late G2XO of London. During the early 1930s Ern built himself a number of short-wave receivers, from designs published in the amateur radio press. His introduction to v.h.f. came in the second world war with the Royal Tank Regiment. After the war he spent some years in East Africa where he used "point-to-point" links, and on his return to the UK, a colleague, G3YHM, invited him to join the Worthing Amateur Radio Club. In May 1972 he took the RAE, and by August, his call, G8GKV, was heard on the air.

In 1927, the 10 -year-old Ern Hoare began building crystal sets, progressing to valve sets, and later to a 30 -line scanning disc television receiver, through which he did get a picture of sorts. Ern was a radar operator with the Royal Artillery during the second world war, using both v.h.f. and microwave systems. Later, he volunteered for the 6th Airborne Division, where he used the famous 38 sets for 3in mortar fire control. He took the RAE in 1962 and was first licensed as G3RZD/T, constructing his own gear for amateur TV. When his call was changed to G8BDJ he used home-brew gear on $2 \mathrm{~m}, 70 \mathrm{~cm}$ and 23 cm .

Due to common interests, both G8GKV and G8BDJ began to inhabit the local hill tops, operating portable stations on 2 m . When 'BDJ began building equipment for 10 GHz , 'GKV was hooked, since when "The Two Erns" have never looked back. Now, with some very impressive home-brew portable stations, they each hold many well deserved awards, and apart from entering the RSGB's "diffz contests, they spend a great deal of time trying out "difficult" paths of microwavelength communications.

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| 7441 | 70 p | 74157 | 70p | 74LS30 | 22p | 9374 | 200 p | 4020 | 100p | LM733 | 100 p | *TDA1022 | 600p | - BF2548 | 35p | *TIP29A |  | 2N2907A |  | 3N: | 10 |  |  |
| 7442 A | 60p | 74159 | 190p | 74LS47 | p | 9601 | 175 | 4021 | 0 p | LM741 | 29p | XR2208 | 400 p | BF257/8 | 32 p | -TIP30A | 55 | 2N3053 | 20 p | 3N204 | 100p | ${ }^{1} 1 \mathrm{~A} 50 \mathrm{~V}$ | 21 p |
| 7443 | 12 p | 74180 | 100p | 74LS55 | 30p | 9603 | 60 p | 4022 | 100 p | LM747 | 70p | XR2207 $\times \times R 2216$ | 600p | BF259 | 36 p | *TiP30C | 60 p | 2N3054 | 65p | 40290 | 250p | *1a 100 V | 22p |
| 7444 | 112p | 74161 | 100p | 74LS 73 | 50 p | INTERF | C | 4023 | 22 p | LM748 | 35p | -XR2216 | 675p | -BFR39 | 30 p | TIP31a | 58 p | 2N3055 | 48p | 4036 | 40 a | -1A 400V |  |
| 7445 | $100 p$ | 74162 | 100 p | 74LS74 | 50 p | NKTER | E | 4024 | ${ }_{20 \mathrm{p}}^{50}$ | LM3800 | 70p | XR×240 | 400 p | -BFR40 | 30 p | TiP3ic | 62 p | 2N3442 | 140p | 40361/2 | 45p | 2A 50V | 30p |
| 7446 A | 93p | 7463 | 100p | 74LS75 | 50p |  | 100p | 4025 | 200 | LM3911 | 130p | ZN414 | 90 p | "BFR41 | 30 p | TIP32A | ${ }^{88}$ | 2N3553 | 240p | 40364 | 120p | "2A 100V | 35p |
| 7447 A | 70p | 74164 | 120p | 74LS83 | 110p | MC1489 | 100 p | 4026 | 130 p 50 | LM4136 | 120p | ZN424E | 135 p | -BFR79 | 30 p | TIP32C | $82 p$ | -2N3565 | 30 p | 40408 | $70 p$ | *2A 400V |  |
| 7448 | 80 | 74165 | 130p |  | $100 p$ $40 p$ | ${ }^{7} 5107$ | 160 | 4027 4028 | 50p | -MC1310P | 150p | ZN425E | 400 p | -BFR80 | 330 p | TIP33A | 990p | *2N3643/4 | 48p | 40409 | 65 p | -3A 200 V |  |
| 7450 | 17 p | 74166 74167 | 1400p | 74LS 90 | 60 p | 75182 | 230p | 4029 | 100 p | MC1458 | 55p | ZN1034E | 200p | - BFR81 | 30 p | TIP33C | 114p | -2N3702/3 |  |  |  | -3A 600 V |  |
| 7453 | 17p | 74170 | 240p | 74LS93 | 60p | 7545 | 120p | 4030 | 55p | MC1495 |  |  | 00 | BFX29 | 30 p | TIP34A |  |  |  | 405 | 370 | * 4 A 400 |  |
| 7454 | 17p | 74172 | 720p | 74LS107 | 45p | 754591/2 | 72p | 31 | 200 p | VOLT |  | ATORS |  |  |  |  |  | *2N3708/9 | 12p | 40585 | 105p | 6A 50V |  |
| 7460 | 17p | 74173 | 120p | 74LS112 | ${ }^{100 p}$ | 75491/2 | 96p | 4033 | 180 D | Fixed |  |  |  | BFX86/7 | 30 p | TIP35 | 29 | 2N3773 | 300p | 40603 | 58 p | 6A 100 V |  |
| 7470 | $36 p$ | 74174 | 93 p | 74LS123 | ${ }^{750} \mathrm{p}$ | C-Mos | 1.C.s | 4034 | ${ }_{110}$ |  |  |  |  | BFX38 | 30p | TIP36A | 270p | *2N3819 | 25p | 40673 | 900 | 6 A 400 V |  |
| 7472 | 30 p | 74175 | $85 p$ | 74LS132 | 900p | 2 | 25p |  |  | 5V 780b | 90p | 5 V 7905 | 100 p | BFW10 | 90 p | TIP36C | 340 | - ${ }^{\text {N }} 3820$ | 50p | 40841 | 90 p | 10 A 400 V |  |
| 7473 | 34 p | 76 | 90p | 744LS ${ }^{\text {743 }}$ |  | $74 \mathrm{CO2}$ 74 CO | 25p | 4041 | ${ }_{80} 100 \mathrm{p}$ | 12v 7812 | 90p | $12 \vee 7912$ | 100 p | BFY50 | 22 p | TiP41A | 65 | 2N3823 | 70 | $40871 / 2$ | 90 p | 25 A 400 V |  |
| 74 74 | 30 p | 74178 | 160p | 74LS139 | p | $74 \mathrm{CO8}$ | 27p | 4042 | 80 p | $15 \vee 7815$ | 90p | 15 V 7915 | 100p |  |  |  |  |  |  |  |  |  |  |
| 7476 | 35p | 74180 | 93p | 74LS151 | 100p | 74 C 10 | 27 p | 4043 | 90 p | 18 V 7918 |  | 18 V |  |  | T | TE |  | ms | 8 | XCE | T | $k$ |  |
| 7480 | 50p | 74181 | 200p | 74LS153 | 60p | $74{ }^{\text {c14 }}$ | 90 p | 4044 | 90 p | 24.7824 |  | 24.7924 |  |  |  |  |  |  |  |  |  |  |  |
| 7481 | 100p | 74182 | 90 p | 74LS157 | 60p | 74 C 20 | $27 p$ | 4046 | 110p | 100 mA | -92 | 100 mA T | O-92 |  |  |  |  |  |  |  |  |  |  |
| 74 | 84 p | 741 | 150 |  |  | $74 \mathrm{C30}$ | 27 p | 4048 | 55p | 5V 78L05 |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 7483 A | 90 p | 74185 74186 | ${ }^{150 p}$ | 74LS164 | 100p | 74 C 42 | 110 p | 4049 | 32 p | 12 V 78L15 |  | 15 V 79L15 | op |  |  |  |  |  |  |  |  |  |  |
| 7485 | 140p | 74190 | 100p | 74LS162 | 140 p | 74.48 | ${ }^{250} 9$ | 4050 | ${ }^{49} \mathrm{p}$ | OTHER 8 |  |  |  | psp |  | AT at |  |  |  |  |  |  |  |
| 7486 | 34p | 7494 | 100p | 74LS163 | 100p | $74 \mathrm{C73}$ | 75 p | 4051 | 80 c | LM309K | 135p | TBA625B | 120p |  |  |  |  |  |  |  |  |  |  |
| 7489 | 210p | 8492 | 100p | 74 LS164 | ${ }^{120 p}$ | $74 \mathrm{C74}$ | ${ }^{700}$ | 4052 | 80 | LM317T | 20.9 | TL430 | 65 p |  |  |  |  |  |  |  |  |  |  |
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a minature won with the element enclosed trist in a ceramic shaft then in a staïniess steel Vilually leak free Only 7 ，＂ long，Filled with a $3 / 32$ bit $14-37$ nclusive of vat and 5 other bits avaliable from wown do 364

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Model SK4 Kit


## Now heat to aly level What 3


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## Antex TCSCI Soldering Stafon：

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## Finishing it off

## Give your project a really professional appearance

A growing number of home constructors seem to be aiming to finish their projects in a more professional manner. There is a growing interest in using better looking and better finished cases and cabinets and the days of the old tobacco tin seem, fortunately, to have passed.

The choice of cases available to the home constructor has improved in recent months and several companies have made some of their professional cases available on the amateur market.

In most projects presented in Practical Wireless the decision as to the type and style of case to be used has already been taken. A lot of thought goes into the choice of case for each project, bearing in mind such parameters as price, the use to which the project will be put, availability and styling.

## Simple Boxes

Some simple projects can be adequately housed in a simple plastic box if electrical screening is not required, or in the traditional diecast aluminium box if it is needed. Plastic boxes have been developed by several makers into quite sophisticated units offering many advantages for the amateur user. They are relatively inexpensive and can be easily drilled for controls using only simple hand tools, although some degree of care is needed if the smooth polished surfaces are not to be marked. A wide range of sizes and styles are available and several types have mounting facilities for printed circuit boards moulded into the box, simplifying construction.

The diecast box in its simplest form has been around for several decades andthere are probably more pieces of electronic equipment built into one of these
than into any other type, with the possible exception of the infamous tobacco tin. It has several disadvantages for the amateur, being quite difficult to work on with simple hand tools and having a rather indifferent appearance in the 'as bought' state. Versions are available with a respectable gloss painted finish, but these are very expensive and unless the electrical screening properties of the diecast box are really needed then the plastic versions are better, and cheaper.

The plain simple aluminium box with removable lid offers a stark and cheap housing without any pretensions as to style or elegance.

## Test Gear

When it comes to projects such as test equipment, domestic audio or amateur radio projects, some respectable form of housing becomes essential.

Test equipment should be built to be used, and must offer reliability and confidence and it cannot really give of its best housed in a piece of bent aluminium. The case design chosen must be mechanically robust, easily worked and allow the controls to be ergonomically positioned. If the instrument is intended to be used in the workshop then allowances should be made for maximum utilisation of bench or shelf space. Take a look at what the commercial instrument makers use and don't be ashamed to take the best ideas from all of them. Don't be tempted to use a case that is so small that the controls have to be cramped together making the instrument awkward to use. You should be proud of the finished instrument and not be ashamed to leave it permanently on the workshop. bench where it will get maximum use.


Audio equipment tends to be ratherfashion conscious and so the cases and cabinets used for amplifiers and tuners change styles rapidly. Because this type of project is on show to anyone who enters your home it is very important to ensure that the workmanship put into the cabinet and front panel of any audio project is of the very best. Probably the easiest way to achieve this is to build from a complete kit where the cabinet is provided. However, if you are designing your own circuits or insist on making the entire unit yourself then there are a few recent additions to several makers' case ranges which would lend themselves to audio equipment.

For the amateur radio enthusiast the need is for a functional but still attractive case at a respectable price and it also probably needs to be all metal as well.

## Metalwork

Although very respectable work can be turned out on the kitchen table using nothing more than a pair of scissors, an old file and a simple hand drill if you have the patience, it is very much easier with a few basic tools.

The home constructor needs to be able to drill holes in metal or plastic panels, file rectangular holes for meters or plugs, cut pieces of metal to size and shape and bend them into simple shapes such as brackets. To perform these operations a few basic tools are essential and you must learn how to use them properly.

Some means of holding the work is essential if painted or polished surfaces are not to be scratched.

Ensure that your cutting tools and drills are all kept sharp and in tip-top condition. Blunt and rusty tools do not

Drilling sizes

| Thread | Clearance | Tapping |
| :---: | :---: | :---: |
| OBA | 6.10 mm | 5.10 mm |
| 2BA | 4.80 mm | 4.00 mm |
| 4BA | 3.70 mm | 3.00 mm |
| 5BA | 3.30 mm | 2.65 mm |
| 6BA | 2.85 mm | 2.30 mm |
| 8 BA | 2.25 mm | 1.80 mm |
| 10BA | 1.80 mm | 1.40 mm |
| 12 BA | 1.40 mm | 1.05 mm |

A professional look can be achieved with the PW Front Panel Overlay System. This picture shows the overlay for the PW Purbeck oscilloscope. The overlay can be used over a plain white card panel or over a coloured one if preferred. A thin Perspex sheet can be fitted over the film to protect it and hold it firmly in place
help to make a good job. Mark out the positions of the holes and cut-outs accurately, paying particular attention to ensuring that groups of holes are correctly positioned relative to each other. Centre-pop each hole before drilling a small diameter pilot hole first then following up with the correct size drill. Rectangular cut-outs are made by drilling holes at each corner and using an Abrafile fitted in the hacksaw frame cut along the four sides. Use a suitable file to finish the cut-out to size.

Wherever possible try to use components that have some sort of bezel to cover up the holes. This gives you much more leeway with your metalwork.


Collet type knobs provide a firm fastening together with ease of positioning on the shaft. These are Sifam knobs and the various component parts can be seen


## Front Panels

Lettering on the front panel can make or mar a project. Although press down letters and numerals can be used with good effect they are not as easy to use properly as is widely imagined and unless they are carefully fixed with special varnishes they tend to rub off quickly in use.

A system to enable constructors to produce professional looking hard wearing front panels for many PW projects has been evolved.

This uses a photographically produced transparent film of the front panel which can be carefully cut out and placed over the main panel. A thin sheet of Perspex can be placed over the film to hold it flat, the "sandwich" being held in place by the controls.

A further stage in the production of front panels is to use the film overlay as a photographic master to make a thin metal panel which is then stuck onto the main front panel. This method produces superb panels but does require the use of ultraviolet lamps, specially sensitised metal sheet and the appropriate chemicals.

## Knobs

To complement the front panel design you should choose suitable knobs, bearing in mind the use to which the unit is to be put. For test equipment where ease of use is vital it is difficult to better a collet type of knob. These fit tightly onto the control shaft by a simple collet action and do not
require a flat on the shaft for a grub screw.

For audio units more fashionable knobs can be chosen but remember that good knobs cost more than poor ones.

## Wire sizes

| s.w.g. |  | dia. mm | s.w.g. |  | dia. mm |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 10 | 3.251 | 30 | 0.315 |  |  |
| 12 | 2.642 | 32 | 0.274 |  |  |
| 14 | 2.032 | 34 | 0.234 |  |  |
| 16 | 1.626 | 36 | 0.193 |  |  |
| 18 | 1.219 | 38 | 0.152 |  |  |
| 20 | 0.914 | 40 | 0.122 |  |  |
| 22 | 0.711 | 42 | 0.102 |  |  |
| 24 | 0.559 | 44 | 0.081 |  |  |
| 26 | 0.457 | 46 | 0.061 |  |  |
| 28 | 0.376 | 48 | 0.041. |  |  |

## Basic tool kit

Hand drill with $\frac{3}{8}$ inch chuck
Drills $1 \mathrm{~mm}, \frac{1}{16}, \frac{1}{8}, \frac{1}{4}, \frac{3}{8}$ inch diameter, countersink bit
Junior hacksaw
Abrafiles (to fit hacksaw)
Centrepunch
Small ballpein hammer
Small files including round, halfround and three square
Steel rule, 12 inches tong
Small engineer's square
Scalpel and blades (also use as scriber for marking out)
Small vice
Screwdrivers
Pliers, small and large
Sidecutters
BA spanners
Soldering irons $15 \mathrm{~W}, 30 \mathrm{~W}$

# The mystic art of soldering 

## Assemble your projects correctly

The absolute success of any electronic or wireless project lies with the workmanship put into the construction. This applies especially to the quality of the soldering. One dry joint and the entire project is likely to fail to work, causing despair to the beginner.

It is simply not sufficient to assume that a project can be thrown together anyhow, the solder blobbed onto the joints, the power switched on and it will work. With modern circuits and exotic components this is just not the case. The would be wireless constructor should aim to master the various constructional techniques used.

The basic art which must be perfected before starting any constructional project is the technique of soldering.


## Choice of Iron

Let us discuss the choice of iron first as this has a direct bearing on how easily you can make a given joint. The only purpose of the iron is to heat the metals to be soldered to the correct temperature so that the solder flows properly. It is not intended to be used as a means of carrying the solder to the joint or as a tool for bending lead ends over. The iron should ideally be matched to the joint being made so that the temperature is neither too hot nor too cold. If the temperature is too hot it is probable that the more delicate components might be damaged while if it is too cold the solder will not melt properly and the joint will be dry.

Soldering irons are available in a wide variety of sizes and shapes. The first figure to be quoted when talking about irons is the wattage of the element and these vary from about 10 W up to about 240W-ideal for soldering the seams on your car radiator. For our purposes a 15/17W miniature iron with an iron plated copper bit of about 2.3 mm diameter is ideal for modern p.c.b. work

Top left is a badly made soldered joint on a p.c.b. probably made with an iron that was cold and dirty. Too much solder has been used and has bridged across the three pads. Below it is the same area of the board after the excess solder had been removed and the joint made with a clean hot iron. Top right shows a solder bridge across two adjacent pads. Below are the same joints properly made. The heading picture shows a Stiron 75W iron and Antex 30W and 17W miniature irons
and will cope with most joints to be found on PW project boards.

However where large diameter earth wires, metal screens and components with heavy gauge wires are met with then such a small iron will not be capable of supplying enough heat to bring the metals up to soldering temperature rapidly enough and a poor joint will be the result. For this type of work and also for older projects such as those using valves a $25 / 30 \mathrm{~W}$ iron with a bit of around 3.2 mm diameter is needed.

If you want to go in for some really heavy brass metal chassis bashing then you will need at least a 60W iron with a large bit and you might even find that a small propane gas blowtorch is better.

## Preparation

When preparing the components for soldering ensure that the leads are clean and, if necessary, tin them by applying the hot iron to the leads together with a resin cored solder so that the lead is completely covered by a thin layer of solder.

The leads should then be carefully bent so that the component fits into place without any strain on its leads. Place the hot iron onto the joint and immediately apply resin cored solder to the joint. The solder should flow easily and quickly to completely cover the joint, when the solder and iron can be removed and the joint allowed to cool. It is very important that the components are not allowed to move during the cooling down period as this will affect the quality of the joint.

When it has cooled down and the solder solidified the joint can be inspected. The solder should completely surround the joint and be bright and shiny. If the joint is dull or crystalline in appearance the joint is "dry" and must be remade. Likewise any small blowholes or areas where the solder has not "wetted" the leads means that the joint is suspect.

One very common fault is to use far too much solder on each joint. Only enough solder should be applied to the joint to completely cover it. An excess of solder is not only untidy, and could cover a bad joint but it is also expensive.

If you find that you have to remove a component from a p.c.b. at any time this can be easily done by using a de-soldering


These two pictures show a TV game p.c.b. as built by a novice (above) and after rebuilding correctly (below). Note how the components are poorly formed and placed in the upper picture with the transformer hanging on one small screw. All components should be neatly formed and positioned with great care and precision to obtain maximum reliability

braid. This braid is placed on the joint and a clean iron applied. The solder melts and is literally sucked into the braid by capilliary action. A clean piece is used for each joint, of course.

The soldering iron bit must be kept clean and this is best achieved by using a damp sponge and wiping the hot bit with it after each joint is made. Several stands are available to hold the iron when it is not in use and these usually incorporate the sponge in their bases. Remember to
keep the sponge damp though.
Never use any other solder than a resin cored variety which has been specially formulated for electrical and electronic work. Fluxes such as Baker's Fluid or other highly active varieties must never be allowed to get anywhere near an electronic component, the results will definitely prove disastrous for the component as well as shortening the life of your soldering iron bit. Leave these corrosive fluxes to the plumbers.

# Identifying components <br> The resistor and capacitor colour code explained 

Although there are some minor differences in specific capacitor ratings, both resistors and capacitors follow the same colour code. In the case of resistors, where the colour code is in use, it refers to a nominal value in ohms, the unit of resistance. There are of course other types of resistor which do not utilise a colour code, such as wirewound types (high current or high stability) and some which simply have the resistance value written as a figure on the resistor body itself, but the vast majority of carbon, metal oxide, and thick film types use the colour code.

The way in which the code works is very simple, the colours being read off from one end of the resistor to the other, beginning at the end where the colours are concentrated. The first ring of colour indicates the first digit, the second ring the second digit, and the third indicates the multiplier or number of zeros in use. The fourth colour indicates the tolerance over the stated range, brown indicating $1 \%$, red $2 \%$, gold $5 \%$, and silver $10 \%$.

A typical resistor might read yellow. purple (or violet if you prefer), red, and gold. This indicates 47000 hms , variously written as $4.7 \mathrm{k}, 4.7 \mathrm{k} \Omega$, or 4 k 7 , and a tolerance of $5 \%$ over that range. It is important to appreciate that the third colour (the multiplier) actually denotes the number of zeros, thus a $47 \Omega$ resistor would appear' as yellow, purple, black, with a tolerance band following, indicating that there are no zeros in the multiplier.

In general, the colour code is restricted to carbon or metal oxide types, and these resistors will of course be suitable for all types of circuit application. On the other hand, wirewound types, which normally have their value stamped or printed on the body of the resistor, will possess inductive properties, which may render them unsuitable for r.f. circuits.

The stability and tolerance of a particular resistor is to a great extent dictated by the materials of which it is made, as is the inherent electrical noise
which it produces. High stability items are typically constructed from carbon film ( $5 \%$ tolerance) or metal oxide ( $2 \%$ tolerance) cermet ("thick film") also $2 \%$ tolerance, but precision items are of course wirewound, giving a rated tolerance of $0.1 \%$.

In the case of power dissipation, wirewound forms are inevitably the favoured construction, giving a dissipation in specific items up to about 50 W commercially, but once again, fairly high levels of dissipation can be managed with moulded carbon compound types, up to about 2 W .

Where low noise levels are concerned, metal oxide and thick film types are used.

## Capacitor Markings

Although the colour code is the same as that for resistors, it is generally


## Table of Multipliers

> (referring to all electrical functions-current, voltage, resistance, inductance, frequency, power, and time)
> $\mathrm{p}=$ pico $=\times 10^{-12}$
> $\mathrm{n}=$ nano $=\times 10^{-9}$
> $\mu=$ micro $=\times 10^{-6}$
> $\mathrm{~m}=\mathrm{milli}=\times 10^{-3}$
> $\mathrm{k}=$ kilo $=\times 10^{3}$
> $\mathrm{M}=\mathrm{mega}=\times 10^{6}$
> $\mathrm{G}=$ giga $=\times 10^{9}$
restricted to ceramic disc, tantalum bead, and some moulded or dipped types. In the case of capacitors, there is some extra information necessary, notably the working voltage. The general code is read in the same way as with resistors, beginning at one end and moving towards the wire ends, the last line of colour indicating the working voltage. In the most common form, the polyester dipped type, red $=$ 250 V and yellow 400 V . In these types a white tolerance band indicates $10 \%$ and black $20 \%$.

Caution must be exercised where colour bands are similar to the colour of the capacitor body, as for example, a 2200 pF (2-2nF or 2 n 2 ) polyester dipped type might well appear as a totally red item, made up from red $=2$ (first digit), red $=2$ (second digit or tens), and red $=$ 2 (multiplier or number of zeros).

Tantalums are usually low voltage types and the code, again, is read from the free end, away from the wire ends. The first colour is the first digit, the second the tens in the figure, the next the multiplier, and the final colour the working voltage. These voltage bands follow a colour code, black indicating 10 V , yellow 6.3 V , green 16 V , blue 20 V , grey 25 V . white 30 V and salmon pink 35 V .

## Applications

The non-electrolytic categories, such as silvered mica, ceramic, paper, polyester and polycarbonate are highly suitable in all signal processing applications, both audio and radio, with the principal restriction being each type's working voltage and temperature capabilities. Ceramics are normally used in situations where their good temperature characteristics are an advantage, such as tuned circuits. Polyester, paper and polycarbonates are universally useful in high voltage situations where good insulation resistance is required such as pulse circuitry in television, and general signal and power stages in radio and audio circuits. In the main, coupling in transistor a.f. stages etc. utilises specific polyester items, due to their smaller physical size and good tolerance.

Electrolytics are used where high capacitance values are needed, and in the de-coupling of power supplies, where hum and ripple may be a problem.

# Component buying 

## How and where to purchase those elusive parts


#### Abstract

"Where can I buy the components for the PW 'Whathaveyou' project?" is one of the commonest queries received. This article will try to explain the pitfalls of purchasing electronic components, and show how to overcome them.

Unless you can buy a complete kit for a project it is quite probable that you will have to undertake some detective work to uncover the relevant suppliers.

The first thing that any serious home constructor needs is a library of catalogues from various suppliers of components. A run through the adverts in PW will indicate who produces a catalogue and you really cannot have too many of them. Using your library you can select the appropriate supplier for most components. You can compare prices for various components to try to obtain the best deal if this sort of exercise takes your fancy.


## Exotic projects

Consulting your favourite supplier's catalogue is all right for the common or garden components such as resistors, capacitors and semiconductors, but what about the special items often specified for the more exotic projects?

Often these are available only from one source and where this is so the source is given in the components list or in the text of the article describing the project.

One name that keeps appearing in component lists is that of RS Components. This company is a component distributor who produce a comprehensive catalogue widely used in industry. All the components sold by RS are "own brand" and they only sell to bona fide trade customers. Because of their rapid service and well-produced technical catalogue,
their components are used widely by authors when preparing projects, and the staff of PW also use them for similar reasons.

This leads to complications when a components list calls for an RS Components part number as many readers will find it difficult to obtain them. One way out is to befriend your local radio and TV repair shop who will certainly have an account with RS. It must be remembered that the prices given in their catalogue are not retail prices, and do not include VAT either, so be prepared to have to pay around $50 \%$ above catalogue price.

Up until recently RS Components' sister company Doram supplied to the amateur market but this service has been withdrawn. However several of the regular advertisers in $P W$ will obtain any RS Components part on request.

By the way, try and persuade your friendly repairman to let you have an old issue of RS Components' catalogue, it is a mine of valuable information on components, giving such information as sizes, connections and specifications.

## Cases and cabinets

When it comes to the cabinet or case to fit the project, these are usually available from several advertisers, most of whom carry a range from two or three manufacturers. In most projects the actual case style is a matter of personal choice and the one used by the author need not be capied. However, some projects are built using specially designed cases or utilise some particular feature of a case, and then it is important that only that case is used.

If you are the type who enjoys making your own chassis and cases then you will
need to find a supplier of raw materials. This is not always easy and it pays in this case to have a copy of Whiston's Catalogue (K. R. Whiston Ltd., New Mills, Stockport SK12 4PT). This lists small quantity sizes of aluminium, brass, steel, etc., and is available for an s.a.e.

Printed circuit boards for PW projects should be obtainable from advertisers, but, individual readers can, if they have the equipment, make their own boards from the copper track patterns given fullsize in the magazine. Another way is to take the track pattern to your local printed circuit board company and get them to make you a board. This is likely to be expensive as they will have.to make you a photographic master from the drawing first (note that this must only be made for you and boards must not be made for resale by this method without obtaining permission and paying royalties to the copyright holder). You can find the address of p.c.b. makers in your local "Yellow Pages", which can also prove useful in locating component stockists, metal stockists, etc.

## Semiconductors

When buying semiconductors for a project you will have to make up your own mind as to whether you risk using cheap unmarked types or spend out more money for guaranteed ones. Most suppliers will honour the maker's warranty and change faulty i.c.s or transistors so long as they have not been soldered into a circuit. With c.m.o.s. types, do not be tempted to take them out of their packaging to look at them. They are very prone to damage from static which builds up on all humans these days. Leave these i.c.s firmly in their conducting foam or silver foil wrappers until you are ready to insert them into their sockets. If you are supplied with c.m.o.s. not protectively wrapped in foil, conducting foam or a special housing. get in touch with the supplier immediately. If possible it is best to buy such components only from suppliers who are recognised or franchised by the manufacturer.

Finally, if you have exhausted all the above possibilities and feel that you simply must ask the magazine staff, please enclose a s.a.e. and don't expect miracles, and please, oh please, only after you really have tried yourself!

# Handling C.m.o.s. 

## Avoid killing your i.c.s with static

Unlike t.t.l. (transistor-transistor logic), c.m.o.s. (which stands for "complementary metal oxide semiconductor") devices are very prone to damage if handled or fitted into circuits without due precautions having been taken.

It is true that one or two digital c.m.o.s. devices are fairly robust, and can usually only be destroyed by reversing polarity of the supplies or by feeding a high signal at low impedance into the inputs while the device is without a supply. but general c.m.o.s. "chips" for use in the audio, video, music and radio fields are highly susceptible to static charges and extremes of temperature.

A great deal has already been written on somewhat amusing lines concerning the wearing of nylon clothing while working on such devices, but the basic facts are quite simple and need no "scripting" in order to emphasise the requirements.

1) Unused inputs should always be tied to ground or positive supply. depending upon requirements.
2) Polarity of supplies must be carefully checked before any connections are made.


As it should be, neat and tidy.
3) Low impedance sources, including charged electrolytic capacitors, must not be connected directly to the input terminals, especially in the case of logic "chips". It is essential to use a surge limiting resistor in such cases of at least $1000 \Omega$.
4) Input terminals must not be allowed to "float" and must be, like unused inputs, tied to ground or positive via a high resistance.
5) Where devices are supplied in packages of conductive foam, conducting foil, or the specialised conducting tubes which are becoming typical of c.m.o.s. packaging, they should not be removed until the very last moment in order to avoid the effects of any local static charges.
6) It is as well to avoid wearing nylon shirts or similar synthetic


Eight


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