

#### T.T.L. 74 I.C's By TEXAS, NATIONAL, I.T.T., FAIRCHILD Etc 14p 14p 14p 14p 40p 20p 20p 20p 20p 20p 7413 7414 7432 7437 7438 7440 7441 7442 7445 7446 7447 7451 7453 7454 7460 7490 7491 7492 7493 7495 7496 74100 74104 74105 74107 74109 74118 74120 35 p 75 p 45 p 40 p 70 p 95 p 40 p 30 p 50 p 90 p 100p 60p 270p 270p 270p 75p 230p 160p 120p 65p 120p 100p 75p 74189 350p 140p 74190 74191 74192 30 p 25 p 7416 7417 7470 7472 74160 74161 90p 90p 74176 74177 100p 140p 74125 74126 50p 65p 65p 86p 85p 75p 15p 15p 90p 90p 140p 140p 74162 74163 74193 74194 120p 120p 100p 74130 74131 74132 74135 130p 100p 65p 100p 7423 7425 7475 7476 30p 30p 74164 74165 74180 74181 100p 200p 74195 1000 7426 7427 7483 7485 85p 100p 74196 74197 100p 74166 74167 125p 325p 74182 74184 75p 150p 74136 74137 74138 80p 100p 100p 185p 185p 74170 74173 200p 74185 74188 74198 **SEMICONDUCTORS** by MULLARD, TEXAS, MOTOROLA, SIEMENS, I.T.T., R.C.A. AA113 AA144 AA217 AC121 AC126 AC127/01 AC128 AC151 AC153K AC176 ACY17 ACY17 AD149 AD161 AD161 AD162 BA138 BA148 BA154 BA157 BA173 BA216 BA316 BA316 BB105A BB110 BC107 BC108 BC108 BC109 14p 13p 14p BF123 BF125 BF125 BF166 BF166 BF173 BF179 BF179 BF180 BF181 BF182 BF182 BF194 BF195 BF195 BF197 BF199 230p 280p 350p 70p 75p 2N1305 2N1306 2N1308 2N1711 50p 8p 30p 20p 42p 47p 125p 45p 25p 40p 15p BC549B BC549C BC169C OC200 ORP12 R1039 TIP29 80 p 70 p 150 p 45 p BC549C BC557 BCY34 BCY70 BCY71 BCY72 BD121 BD123 BD124 BD131 BD132 30 p 19 p 13p 80p 15p 20p 15p 85p 100p 85p 36p 36p AC1277 195 AC1277 195 AC1276 187 AC1276 187 AC128 185 AC128 185 AC128 185 AC151 257 AC153K 495 AC176 205 AC176 205 AC177 355 AC136 495 AC176 205 A BFX29 BFX81 BFX84 BFX85 BFX87 2N2483 2N2906 2N2907 2N39053 2N39054 2N39055 2N3702 2N3702 2N3702 2N3706 2N3711 2N3715 2N3715 2N3716 2N3904 2N3716 2N3904 2N3904 2N3904 2N3904 2N3906 BC182 BC182L BC183L BC183L BC184 BC184 BC205 BC212L BC213L BC214L BC214L BC258 BC293 BC214L BC258 BC293 BC214L BC258 TIP29 TIP29A TIP29C TIP30 TIP30A TIP30B TIP30C 47p 75p 55p 58p 65p 80p MJE340 MPSA06 OA10 TIP42B TIP2955 TIP3055 TIS90 TIS91 85p 70p 55p 25p 25 p BFX88 BFY50 0 447 OA90 OA91 OA202 OC23 OC28 OC35 OC42 OC45 OC71 OC72 OC75 OC81 OC81 OC83 BD132 BD133 BD135 BD136 BD137 BD139 BD140 BD181 BD182 BD207 BD263 BD263 BD263 BD263 BD263 BD762/01 BF120 BF121 BFY51 BFY52 BFY990 BR101 BRY39 BRY56 BSX20 BSY40 BS105 BU105 BU105 BU105 BU105 BU106 BY100 BY126 BY127 BY133 IN914 IN3754 IN4001 IN4002 IN4003 5p 10p 5p 5p TIP31 TIP31 A TIP31B TIP31C TIP32 BC109C BC113 BC113 BC117 BC125 BC126 BC146 BC147 BC148 BC148C BC148C BC157 BC158 BC159 BC167A IN4005 IN4005 IN4006 IN4007 IN4148 TIP32A TIP32B TIP32C BF199 BF200 BF224 BF225 BF241 BF244B 35p 35p 25p 30p TIP33 TIP33A TIP33B TIP33C BC323 BC328 BC328 BC338 BC547 BC547B 60p 13p 12p 12p 13p 12p IS44 2N456A 2N697 2N929 BF257 BF258 BF259 BF274 BF324 30p 30p 25p 50p TIP34 TIP34A TIP34B TIP35 2N930 2N1302 OC84 50 p 225p BZY88 ELECTROLYTIC CAPACITORS / 330/35 25 5p 330/50 5p 470/10 5p 470/25 CA3090AQ RESIS TORS MULLARD POT CORES SKELETON PRE-SET POTEN-TIOMETERS 400M.W. 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A 60	0V	40p	7815	140p	56V 15	
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DACS WS Chrome on Strap
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DABS Series (Right)
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405 ILB



700



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E1 · 00*
E1 · 69*
C1 49
74p*
74p*
E1 · 00*
3
50p*
10p*
15p*
10p*
15p*
10p*
15p*
12p
10p
20p*
10p
14p

DISPLAYS (Red Led)

BC147/8/9 BC167/8/9 BC177/8/9 BC177/8/9 BC192/3/4 A or L BC212/3/4 A or L BCY70/71/72	12p 10p 20p* 10p 14p 20p*
	1 - 50*
MJE3055	85p*
ORP12	60p*
TIP41 or 42A	65p*
TIP2955	65p*
TIP3055	55p*
TIS43	30p*
ZTX107/8/9	20p
2N2646	50p*
3N2904/5	30p*
2N2926 rbo	10p
2N2926YG	15p

2N3053 2N3055 2N3702/3/56 2N3704 2N3819 & FET	24p" 45p* 10p 9p 18p
 2N3820 FET 2N3904/5/6 2N5457 FET 2N5777 OPTO	38p 20p 50p 50p*
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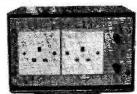
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available separately for both units. 20 + 20 Watts r.m.s. into 8 ohm load. Distortion less than 0-1% 100Hz-10kHz. Frequency response ± 1dB 20 Hz to 20kHz. Hum level virtually nil with volume full on. This is a power amplifier of superb quality incorporating the very latest design features. Professional hi-fi enthusiasts have classed it as fantastic and real value for money. The CCT incorporates a low flux transformer and inputs for disc, tape, tuner, etc.

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8 inch system

8 inch system is designed for use with above amplifiers rated up to 25W t.m.s. per channel at 8  $\Omega$ . May be incorporated in an enclosure 295 x 490 x 295mm (11:5 x 19:3 x 11:5in) approx. external, constructional details of which are given with each bass unit, to provide an overall frequency response of 50Hz to 22Hz. Four-element cross-over, ready constructed on p.c.b. Output leads have push-on receptacles to suit speaker tags. Cross-over frequency is 2.8kHz approx.

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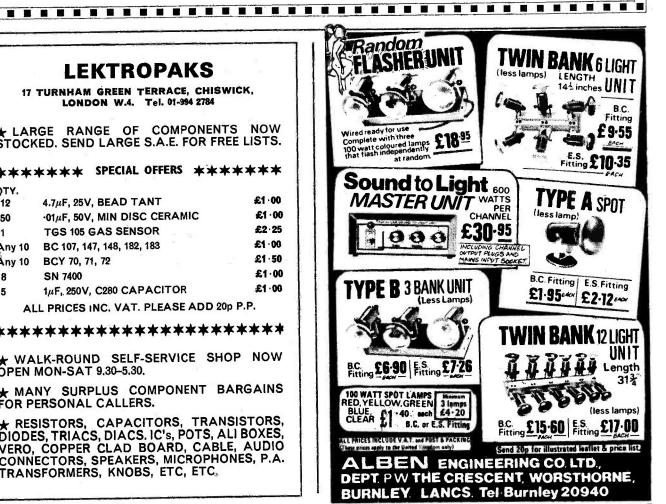
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Practical Wireless, September 1977

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£1700 PER STEREO PAIR + P & P £3.40

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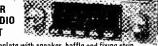
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**EASY TO BUILD** RECORD PLAYER KIT Ideally suited for the constructor who requires a complete stereo unit at a budget price, comprising ready assembled stereo amp. module, Garrard auto/manual deck with cueing device, pre-cut and finished cabinet work, Output 4 watts

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£2750 + p & p £2.50 Size approx 13音"×5½'×6至

Here's the mono unit you need to start off with Gives you a good solid 35 watts rms, 70 watts peak output. Big features include two disc inputs, both for ceramic certridges, tape input and microphone input. Level mixing controls fitted with integral push-pull switches, independent bass and trable controls and master volume.

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(Complete with circuit-diagrams)



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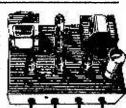
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SPECIFICATIONS INPUT SENSITIVITY 500mV

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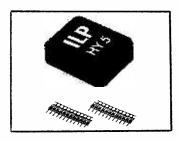
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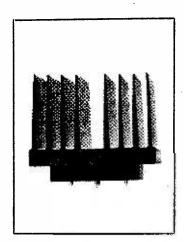
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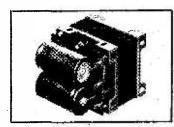
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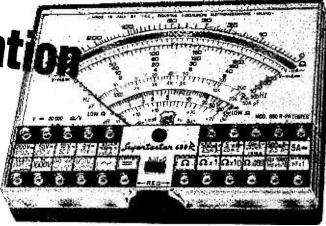
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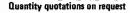
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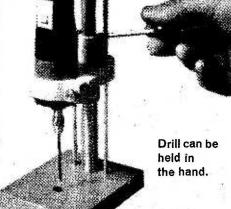
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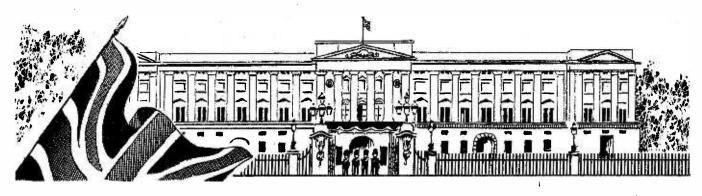


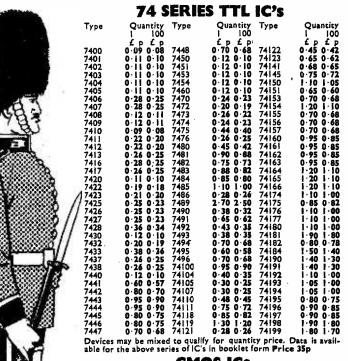
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#### **BINDERS**

Binders, for either the old or the new format, are £2·10 and Indexes are 45p (Inc VAT) and can be obtained from the Post Sales Department, IPC Magazines Ltd., Lavington House, Lavington Street, London SEI 0PF. Remittances with overseas orders for binders should include 60p to cover despatch and postage.

#### **BACK NUMBERS**

We are very glad to announce the re-establishment of a PW Back Numbers Service for our readers. In future back numbers dated from June 1977 only will be available from our Post Sales Department for 65p, which includes postage and packing. Cheques and Postal Orders should be made payable to IPC Magazines Ltd.

Send your orders to:- Post Sales Department, IPC Magazines Ltd., Lavington House, Lavington Street, London SE1 0PF.

# Our Face~lift

ell, here it is! The new and, we hope, more attractive, larger version of **Practical Wireless**, as promised last month. PW has gone through other changes of size in the 45 years that it has been around and when it was first published in 1932 the page size was just the same as our 'new' format! So we seem to have gone through the whole cycle! In the 60's PW was quite small, about 9 × 6in., with correspondingly more pages but it was quite popular since one could roll or fold it up to fit the coat pocket, and I would imagine that there are many readers who would still prefer it that way!

Those that know all about these things have now been able to produce a bigger page area from the same size reel of paper, although the printing costs are marginally higher. We trust that you will like the slightly larger type face, being easier to read. Anyway, if you do favour the new-look **Practical Wireless** do not hesitate to let us know. No doubt the disapprovers will also be in touch but their letters will be equally welcome.

As the compiler of the amateur bands section of our monthly feature "On the Air" I know something of the problems that can be encountered by the very keen listener to the amateur bands who wants to get on and obtain the amateur transmitting licence.

Although many schools, colleges and radio clubs run classes for the Radio Amateur's Examination there are many potential candidates who just cannot reach such places and who must rely on their own studies at home.

This month sees the start of a new series aimed not only at the the person who, perforce, has to be a loner, but all those who aspire to the RAE. Remember, it is unlikely that any two candidates will be starting from precisely the same level. There is the newcomer to electronics, who, by sheer enthusiasm, will spend every possible moment studying both the theoretical and practical side of our hobby, and there is the electronic engineer, with degrees after his name, who needs to swot up on certain aspects of the subject before sending in the rather exhorbitant entrance fee. Today, it behoves every candidate for the RAE to be really sure that he is prepared for the exam. Whatever his capabilities may be on the technical side he must still swot up on the legislation governing the issue of an amateur licence. Part 1 of the exam is not in the least difficult but a failure in this part means a failure of the whole paper, even if you are a veritable genius at electronics!

So we hope that our RAE course will help readers of all technical abilities and, probably, many other readers too, who may never have heard of the RAE!

Eric Dowdeswell Assistant Editor

#### PLEASE NOTE

We do not operate a Technical Query Service except on matters concerning constructional articles published in PW. We do not supply service sheets or information on commercial radios, TV's or electronic equipment.

All queries must be accompanied by a stamped self-addressed envelope otherwise a reply cannot be guaranteed.

# NEWS...

# NEWS.

# NIWS...

#### New catalogue

Bi-Pak has announced the publication of its latest catalogue containing over 120 pages of electronic component listings. There are numerous, albeit rather small, pictures and line drawings. The semiconductor section generally includes pinout details and limited operational parameters. The cost is 50p + 15p postage.

Bi-Pak Semiconductors, The Maltings, 63A High Street, Ware, Herts SG12 9AD.

#### **Audio Fair '77**

Don't forget that the Audio Fair '77 is coming shortly and promises to be the most comprehensive ever staged. Displays will range from the most sophisticated BBC/IBA/GPO Data transmission modes to the more mundane HiFi equipment, TV games and home movie apparatus.

The show which is at London's Olympia will open from 2pm-9pm on September 13th, from 10am to 9pm on September 14th, 15th, 16th and 17th, and from 10am to 8pm on Sunday, September 18th. Admission will be 70p for adults, and children under 14, 30p.

#### **New distributor**

Plessey Aerospace has appointed Sasco as sole stockist for their complete range of Plessey miniature relays. Products covered by this agreement include half-crystal-can and full-crystal-can types (sealed and unsealed), hermetically sealed 10A crystal-can versions, and rotary armature Type II relays.

Sasco, PO Box 2000, Crawley, Sussex.

#### New service

Audio Workshops Ltd., have now opened their new showroom at 33 London Road, Southborough, Tunbridge Wells, Kent, where they will be stocking a wide variety of FM aerials, aerial accessories, cables, rotators and electronic car aerials.

The showroom also displays a wide variety of aerial amplifiers from simple low-cost single transistor amplifiers to high quality semi-professional low noise masthead amplifiers.

Full advice as to which aerial or accessory should be purchased to obtain optimum performance will be available from the founder of Audio Workshops, Mr Peter Dunkley. And as a small incentive to go along and see what's available, the firm is offering a high quality 8 element FM aerial at £12.75—for a limited period.

#### A winner

Marshall's have produced a brand new catalogue in a much larger format than before, a bit bigger than this issue of PW. Many new lines include microprocessors and supporting devices, a range of Digital Multi Testers and a pull-out Transistor Guide for the shack wall. Cost is 25p to callers or 35p by post.

A Marshall (London) Ltd., 42 Cricklewood Broadway, London NW2 3ET.

#### Talking show

An exhibition to commemorate Thomas Edison's invention in 1877 of the Talking Machine is to be held in conjunction with the British Institute of Recorded Sound during August. Comprising some 125 exhibits, the

theme of the displays will be the evolution of the phonograph and gramophone from the original hand turned models to today's HiFi. Early original records together with some interesting accessories complete the displays.

Spare parts manufactured by the organisers of the exhibition for collectors will also be available, as will reprints of early catalogues, posters and instruction manuals.

Called "100 years of Recorded Sound" the exhibition will be open to the public from Saturday 13th August to Saturday 27th August (excluding Mondays). Opening times will be 10am until 6pm.

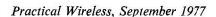
Further information from Goodwin Eve, City of London Phonograph & Gramophone Society, Farm Cottage, High Road, Chipstead, Surrey. Tel: 07375 54827.

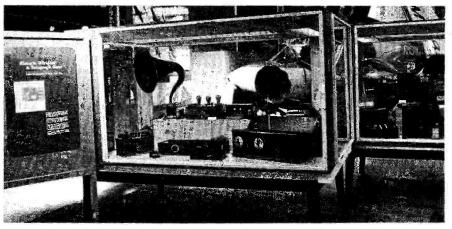
#### Jermyn gen

Jermyn Distribution have now made available the new National Semiconductor MOS/LSI Databook. Containing 400 pages divided into 14 sections, this Databook lists details of electronic organ and TV circuits; A/D converters; communications/CB radio circuits; watches, calculators and keyboard encoders.

In each section, complete characteristics, connection details, function descriptions, and block diagrams are all covered, while the last chapter gives physical dimensions of all National Semiconductor products together with a definition of terms. Priced at £5.95, it can be obtained from Jermyn Distribution, Sevenoaks, Kent.

Part of Ron Ham's early wireless collection which is on display at The Cornwall Aircraft Park, Helston, Cornwall until October 31st. Some of the items on exhibition were featured in Practical Wireless, September 76. On the left hand side of the show case is a copy of the article which was published in The Times newspaper, September 26, 1975 about the collector and his historic pieces. Ron is a very fervent supporter of the new industrial museum that is to be built at Houghton in Sussex and which, he expects, will be the permanent home of his collection that dates back to 1900.





# So you want to pass the R.A.E. (Radio Amateurs' Examination) 🔧

John Thornton Lawrence GW3JGA & Ken McCoy GW8CMY

The passing of the Radio Amateurs' Examination, set by the City and Guilds, requires a certain level of theoretical technical knowledge. Whether one considers that this level is too high or too low is beside the point. The course that follows is intended, with the help of certain external aids, to prepare the reader to pass the examination. It will not teach him all about electronics!

The Radio Amateur is a privileged person indeed, he has access to various parts of the radio frequency spectrum which enable him to talk to other amateurs near to him or in any other part of the world. The Radio Amateur has his own international language, he belongs to a worldwide brotherhood of Radio Amateurs whose "Ham Spirit" takes no account of race, colour, political or social outlook. To ensure that these privileges are not abused, the Licensing Authority (the Home Office) require some test of the would-be amateur's technical competence and knowledge of the radio "Highway Code", before giving him a licence to drive on the frequency motorways of the world, in short, the Radio Amateurs' Examination.

#### ABOUT THE AUTHORS

John Thornton Lawrence, T.Eng. (C.E.L.), F.S.E.R.T., is a Senior Scientific Officer concerned mainly with electronic instrumentation. measurements and calibration. He obtained his ticence GW3JGA in 1953 and his amateur TV Neence GW3JGA/T (later GW6JGA/T) in 1958. He is a Committee member of the British Amateur Television Club and a Founder of the University College of North Wales ARS.

His hobbies include amateur VHF, UHF and IV equipment collecting antique radio gear, particularly valves, wine and beer making and belying the local Scout Group He is married with two sons, 19 and 9

Ken McCoy holds an HNC in Electrical and Electronic Engineering and is concerned with instrumentation and electron microscopy. He is a part-time lecturer for the Radio Amateur's Examination course at the Gwynedd Technical

He obtained his licence GW8CMY in 1967 and his main interest is in building portable 2m equipment. His other hobbies include mountain-eering in the UK and on the Continent, photography and DIY home improvements. He is a team leader in the Ogwen Valley Mountain Rescue Organisation. He is married with a son, 14, and two daughters, 12 and 6.



.... 'writing to the City and Guilds'

Maybe you are one of those whose interest in Amateur Radio has been building up steadily for some time and the point has now been reached where you have to decide whether or not to enter for the R.A.E. If you are dithering on the brink, you may be interested to know that this is the first of a series of articles concerned with preparing for and successfully taking the Radio Amateurs' Examination. It is not intended to be a complete course leading up to the R.A.E. but more of a progressive clarification of some of the problem areas which can be the undoing of many candidates.

The main core of information is best obtained from books specially written for the Radio Amateur and for the R.A.E. in particular and these will be referred to from time to time. Practical work is an important part of preparing for the R.A.E. and suggestions for this will be included where necessary. If by now you have decided to proceed, then please read on.

#### Taking stock

We are very pleased to see that you have now made up your mind to have a go at the R.A.E. Before we do anything else, let's take stock of the situation. You say that you did the usual basic arithmetic in school (but that seems a while ago!) and you naturally feel a bit rusty in that area and with electrical theory, even more so. OK, so some work is definitely required to bring these up to scratch.

How about circuits? You're able to follow some of the simple circuits without much difficulty but find it quite a struggle to explain, in so many words, how they work. Well, perhaps this is just because you are not looking at them in the right way, so more about that later.

Now the practical side. You say that you have built up the odd electronic circuit from PW and have got it working OK, fine! By the way, do you go to the local Amateur Radio Club meetings? No? Pity! You see, it's very helpful if you can do some practical work in connection with the R.A.E. and membership of the local Radio Club is very useful in this respect as it will give you the chance to see amateur radio equipment being used and to obtain some guidance and help when you are building things for yourself. Write to the R.S.G.B. for details of your local club as many clubs are affiliated to the R.S.G.B.

#### **Preparations**

You will need to get some books and information on the examination, past R.A.E. papers and so on. These take a little time to come so I suggest that you order these now. First write to: The City and Guilds of London Institute, 76 Portland Place, London W1N 4AA, for Pamphlet 765 Radio Amateur's Examination 1976-1978, including a set of Question Papers for the last three years. Remember to enclose a cheque or P.O. for 40p. This represents very good value, as usually each set of City and Guilds Question Papers is extra.

The Home Office publishes a useful document about becoming a Radio Amateur and this is available free of charge. Write to: The Home Office, Radio Regulatory Dept., Waterloo Bridge House, Waterloo Road, London SE1 8UA, for leaflet "How to become a Radio Amateur".

Now, moving on to textbooks and booklets, these can be roughly divided into two categories, the essential, and the desirable but not essential. The essentials, which are all published by the Radio Society of Great Britain, can be obtained from your local bookshop (if you want to use up that birthday book token from auntie!) or, cash with order, from: R.S.G.B. Publications (Sales), 35 Doughty Street, London WC1N 2AE.

They are:

"A Guide to Amateur Radio" by Pat Hawker G3VA. £1.35.

"The Radio Amateurs' Examination Manual" by G. L. Benbow G3HA. £1.35.

"Radio Amateurs' Revision Notes" by G. L. Benbow G3HA. 61p.

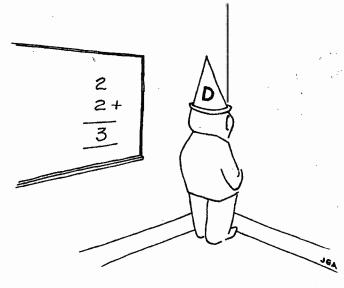
The books in the second category are much more comprehensive and naturally cost more. These are also available from R.S.G.B. Publications (Sales) but as an alternative to immediate purchase, they may be obtained from your local library. They are: Radio Communication Handbook (R.S.G.B.) 5th Edition Vols. 1 and 2, or Radio Amateurs' Handbook 1977 (A.R.R.L.).

#### Getting down to it

The method of studying for the R.A.E. depends on your personal circumstances. The various ways are:

- (a) Evening classes.
- (b) Correspondence course.
- (c) Home study, with or without some tuition.

If you live near a technical college or school which runs evening classes, then this is probably the best way for you to study. A direct approach to the college or school will provide you with full details. Correspondence courses or home study for the R.A.E. require continuous and regular work sessions. Home study requires a lot more personal discipline to get through the more difficult items as you will find it a great temptation to leave them until the end, when it is usually too late!



. . . 'mathematics a bit rusty?'

It is always useful, and particularly so in the case of home study, if you can enlist the help of a local Radio Amateur or suitably qualified friend, who will talk-through any problems you may have.

Here we are then, you have registered at your local evening institute for the R.A.E. course, the books are ordered and you are ready and waiting to get down to the "nitty gritty".

#### Arithmetic

Let's start with arithmetic (we always think that the word arithmetic is less daunting than mathematics). There are going to be some calculation questions in the R.A.E. and if you can do these correctly, you can gain full marks on each question in a matter of minutes. Even fairly straightforward electrical calculations can present very real difficulties if you have not seen the sharp end of a blackboard pointer since you left school, so the purpose of this section is to remind you of the basics involved in these sums and to suggest a few tricks to make them easier to execute and remember.

If you have read through the instruction book of a pocket calculator you will know that there are four "basic arithmetical functions", namely, addition, subtraction, multiplication and division.

Addition and Subtraction—So far as the problems that you will meet in the R.A.E. are concerned, addition and subtraction will take care of themselves, for example, 3+7=10 and 4-2=2. Multiplication and division may be written down in several different ways, so here are a few alternatives.

Multiplication— $A \times B$  can also be written as A.B. or simply AB.

**Division**—A÷B can also be written as  $\frac{A}{B}$ . Thus  $1\div 2$  is the same as  $1_2$  ("one upon two" or "one over two").

Now for the not-so-basic arithmetic functions, reciprocals, squares and square roots.

**Reciprocals**—The reciprocal of A is  $\frac{1}{A}$  or from division  $1 \div A$  or "one upon A", and to put it numerically, the reciprocal of 4 is  $\frac{1}{4}$  and of 100 is  $\frac{1}{100}$ .

This function, together with the squares and square roots, occurs often in resonance and frequency calculations.

Squares—If a number is multiplied by itself the answer is known as the square of the number. For example,  $A \times A$  is written as  $A^2$  or "A squared". Thus  $A^2$  is the square of A. And, to put it numerically, 9 is the square of 3, since  $3 \times 3 = 9$ .

Square Roots—The square root of a number is a figure which, when multiplied by itself, is equal to the original number. For example, the square root of 9 is 3 because when 3 is multiplied by itself, it equals 9, written  $\sqrt{9}=3$  because  $3\times 3=9$ . Similarly,  $\sqrt{25}=5$  (because  $5\times 5=25$ ) and  $\sqrt{4}$  or "Root 4"=2  $(2\times 2=4)$  and so on.

Incidentally, not all square roots are whole numbers. For instance,  $\sqrt{2}$  has been calculated to at least 50 decimal places, but is used most often as 1.41. Finally,  $\sqrt{2} \times \sqrt{2} = 2$  and  $\sqrt{5} \times \sqrt{5} = 5$  and  $\sqrt{25} \times \sqrt{25} = 25$ 

Problems in the actual R.A.E., which involve finding the square roots of numbers, can be dealt with by looking up the value in the square root tables which are provided.

#### Multiples and sub-multiples

Some of the electrical units that are used in radio are inconveniently small for practical purposes and so a multiple of the unit is used. Some others, for example, the Henry and the Farad, are inconveniently large. A one Farad capacitor having a particular dielectric between its plates, might occupy a volume similar to that of a block of flats!

The capacitors and inductors used in radio work have values which are sub-multiples of the Farad and the Henry. You will, therefore, need to be able to recognise multiples and sub-multiples in words and in figures.

Multiples—Multiples are described by a prefix to the unit in question. For example, 1 Megohm = 1,000,000 ohms. A list of multiple prefixes is given below.

Prefix	Le	tter	Index	Form
tera	Т	one million million	1,000,000,000,000	1012
giga	G	one thousand million	1,000,000,000	109
	M	one million	1,000,000	106
kilo	k	one thousand	1,000	103
hecto	h	one hundred	100	102

Sub-multiples—Sub-multiples are described by a letter prefix in a similar fashion. For example, 0.000,001 Farad=1 microfarad=1 $\mu$ F. A list of submultiple prefixes is given below.

Prefix	Le	ter	· Index	Form
deci centi milli micro nano pico (or	d c m μ n p	one tenth one hundredth one-thousandth one millionth one thousand millionth one million millionth	0·1 0·01 0·001 0·000,001 0·000,000,000 0·000,000,000,001	10-1 10-2 10-3 10-6 10-9 10-12

#### Indices

You will notice from the previous sections that the multiplier and sub-multiplier are written as a figure followed by or preceded by a number of noughts.

This number, when expressed in what is known as "index form", uses a special notation which is best shown by example:

$$mega = 1,000,000$$
 In index form 10<sup>6</sup> giga = 1,000,000,000 In index form 10<sup>9</sup>

In the case of multiples, the index number is equal to the number of noughts following the figure, so that, for example:

2.2 Megohms is written as 
$$2.2 \times 10^6$$
 and 7.3 Gigahertz is written as  $7.3 \times 10^9$  Hertz.

Sub-multiples use the same notation but with a slight difference:

pico = 
$$0.000,000,000,001 = 10^{-12}$$
  
micro =  $0.000,001$  =  $10^{-6}$ 

You will notice that in the case of sub-multiples, the index number is minus and equivalent to the number of noughts plus one following the decimal point. For example:

2.2 picofarads=
$$2.2 \times 10^{-12}$$
 Farads.  
7.3 microhenrys= $7.3 \times 10^{-6}$  Henrys.

It is often desirable to convert from one multiple of an electrical unit to a different multiple and this can be done by moving the decimal point and at the same time changing the index number accordingly. For example:

$$2 \cdot 2 \text{ Megohms} = 2,200,000 \text{ ohms}$$
  
 $= 220,000 \times 10 \text{ ohms}$   
 $= 22,000 \times 10^2$   
 $= 2,200 \times 10^3 \text{ ohms (kilohms)}$   
 $= 220 \times 10^4 \text{ ohms}$   
 $= 22 \times 10^5 \text{ ohms}$   
 $= 2 \cdot 2 \times 10^6 \text{ ohms (Megohms)}$ 

Similarly,

$$\begin{array}{l} 0 \cdot 000,\!000,\!002\, Farads = 0 \cdot 002 \! \times \! 10^{-8}\, Farads \\ = 0 \cdot 002\, microfarads \; (\mu F) \\ = 2 \times 10^{-9}\, Farads \\ = 2\, nanofarads \; (nF) \\ = 2,\!000 \! \times \! 10^{-12}\, Farads \\ = 2,\!000\, picofarads \; (pF) \end{array}$$

Multiplying—When multiplying figures in index form it is only necessary to add the indices. For example:

$$10^4 \times 10^6 = 10^{10}$$
 $10^4 \times 10^{-6} = 10^{-2}$ 
 $10^{-4} \times 10^6 = 10^2$ 
 $10^{-4} \times 10^{-6} = 10^{-10}$ 

Dividing—when dividing figures in index form it is only necessary to subtract the indices. For example:

$$\frac{10^{6}}{10^{4}} = 10^{6} \div 10^{4} = 10^{2}$$

$$\frac{10^{-6}}{10^{-4}} = 10^{-6} \div 10^{-4} = 10^{-2}$$

$$\frac{10^{-6}}{10^{4}} = 10^{-6} \div 10^{4} = 10^{-10}$$
Also
$$\frac{10^{-6}}{10^{4}} = 10^{-6} \times 10^{-4} = 10^{-10} *$$

\* An alternative way to deal with an index quantity that you are dividing by is to change the sign of the index and multiply, by adding the indices.

continued on page 342

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A frequency coverage of approximately 1.7MHz to 15MHz includes many useful bands and is obtained in this receiver by two switch-selected ranges. The tuned circuits comprise the four coils shown in Fig. 1, and a few points should be noted.

As an intermediate frequency of 1.6MHz is used, greater freedom from second channel interference is obtained than with a 470kHz IF. Despite this, it is not possible to avoid such interference with a single tuned circuit, especially at higher frequencies, so a matching tuned RF preselector unit is advisable and this can be added later if wished (Part 2). With the RF preselector connected the RF stage drain is supplied via R1, to make use of the primaries of the coils L1 and L2. C1 is external to the receiver. If it is not intended to add the RF unit, C2 and R1 can be omitted, and pins 9 of L1 and L2 connected to chassis.

#### **Circuit Description**

The 4-pole switch S1a-d selects the appropriate pair of coils. These are tuned by the ganged capacitor VC1/2. Instead of having four pre-set trimmers, TC1 is provided across VC2, and VC3 is a panel control. This provides much more latitude in alignment. VC4 is the oscillator bandspread or fine tuning capacitor, which is useful for precise tuning or listening over a narrow band.

Data for the coils used specify a 960pF padder CP1

for L3, and 340pF padder CP2 for L4 using a 2 x 315pF gang. However, 1000pF and 330pF will be suitable and are more easily obtainable. The larger capacity gang is also a standard item and simply increases coverage slightly.

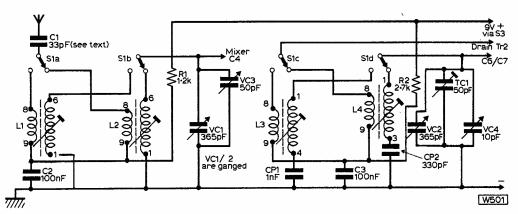
In Fig. 2, Tr1 is the mixer with L1/L2 switched to gate 1 and the oscillator coupled to gate 2 via C6. Tr2 is the oscillator with C7 going to L3/L4 and drain feedback via S1c. As Tr1 is a protected gate device no particular care is needed beyond that generally taken to avoid overheating while soldering.

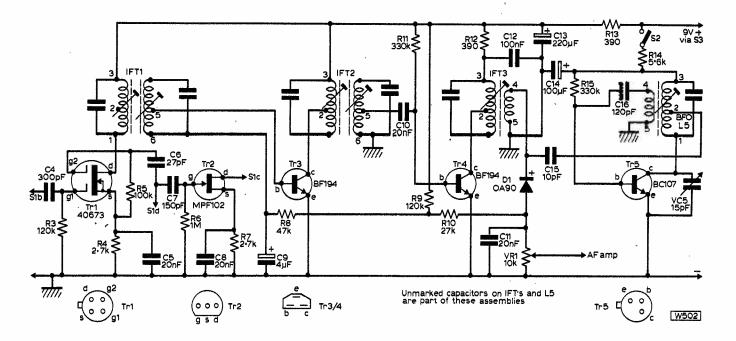
IFT1 couples the mixer output to the first IF amplifier Tr3, which also receives AGC base bias through R8 and R10, from D1. Tr4 is the second IF stage. D1 provides AGC and the audio signal developed across the volume control VR1. When S5 is closed, the BFO oscillator Tr5 is on and it is coupled to D1 by C15. This allows reception of CW (Morse) and SSB (single sideband) signals. The panel control VC5 is for adjusting the BFO frequency. Should the receiver be required for ordinary SW broadcast reception only, then S5, R14, R15, C14, C15, C16 and Tr5 and the associated circuitry may be omitted as this stage is not used for AM reception, but this is not recommended.

This circuit is assembled on a single board, with the exception of controls VR1 and VC5.

The audio amplifier consists of a TBA800 IC, Fig. 3, and provides useful gain and output without too much

Fig. 1: The circuit diagram is shown in three parts. This circuit is of the four-coil switched aerial and mixer input section. Two bands, 1.7 to 5.5 and 5.5 to 15MHz are provided.





drain on the battery. It is intended for use with a  $16\Omega$  speaker but may be used with phones for individual listening.

The audio amplifier is wired on its own board which fits under the chassis. R19 can be two  $2 \cdot 2\Omega$  resistors in parallel if any difficulty arises in obtaining  $1\Omega$ .

#### Metalwork

Flanged "universal chassis" members from Home Radio allow an inexpensive case to be made easily. The panel used is 9 x 4in and the chassis 8 x 4in so that it may fit inside. It is secured lin up, with three bolts. If panel and chassis members are the same length, two corners of the chassis part must be cut away so that it will fit inside the flanges on the panel.

A clearance hole is required for the ganged capacitor which has an integral ball drive. The BFO switch S2 occupies a slot, which can be made by drilling a few small holes and filing. The mixer-IF board has four fixing bolts, and the AF board two bolts. To assure an easy fit, boards and chassis can be drilled together before mounting any components.

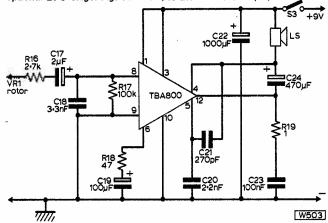
#### Construction

Two 4BA bolts secure the ganged capacitor. Solder a lead to the central frame tag of this before fixing it in place. It is necessary to use washers or other spacers between it and the chassis, and to watch that the bolts do not project very far inside the front of the capacitor. VR1 and other controls can then be fixed to the panel. When placing the coils on the chassis, tighten their fixing nuts with the fingers only. Most of the wiring in Fig. 1 can then be finished, keeping leads reasonably short and direct. If the coil pins are dull clean them before soldering. It is necessary to avoid lengthy heating of the pins and to see that they are not displaced.

Fig. 4 will assist in wiring. One tag of S6 is used as a point for R1, R2, S2 and R13 positive connections. The AF amplifier positive circuit is made at S2, under the chassis. VC4 is directly under the ganged capacitor, so a lead passes down through a hole to it from VC2. Fig. 5 will allow the switch wiring to be checked.

Fig. 6 shows both sides of the mixer-IF board. It should first be drilled for the IFTs, including the

Fig. 2: above, shows the mixer stage Tr1 and oscillator Tr2. Tr3 and 4 are the IF amplifier stages feeding the diode detector D1.Tr5 is the optional BFO stage. Fig. 3: below, is the IC audio amplifier.



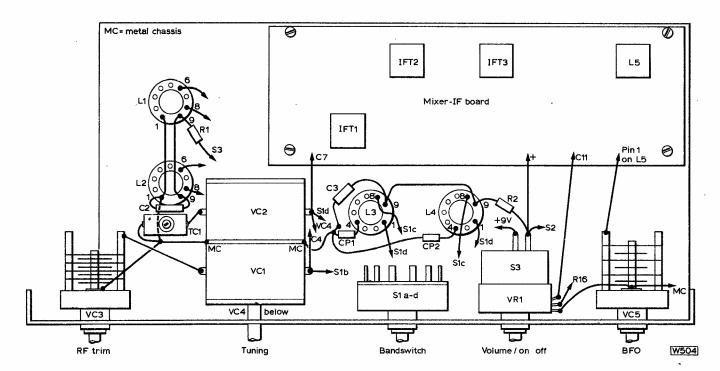
centre holes so that the lower cores can be reached. Each of the points MC is a tag secured with a  $^{1}2$ in 6BA bolt. Later, extra nuts allow the board to be mounted by these bolts, with clearance for wiring. Sleeving should be put on the positive circuit wiring to pins 3 of the IFTs, but is not needed elsewhere.

Provide coloured insulated leads for all external connections. These can be yellow for VC1 (S1b), black for VC2 (S1d), brown for S1c, red for positive, pink for S2, and yellow-green for VR1. When the board is fixed with lock nuts, cut these leads as required and solder them to the capacitor and other components.

The audio board is wired as in Fig. 7, and is mounted in the same manner as described for the other board. Connect R16 to VR1 wiper and positive to the positive line at S2. The board is placed under the chassis, at the right. It is important to note that both sides of the speaker circuit are insulated from the metal and negative line.

#### IF Alignment

As the IFTs are pre-tuned by the manufacturer they should not be adjusted until reasonable reception is being obtained. They can then be set for maximum volume, by rotating the cores with a proper trimming tool. Choose a weak, steady signal for this



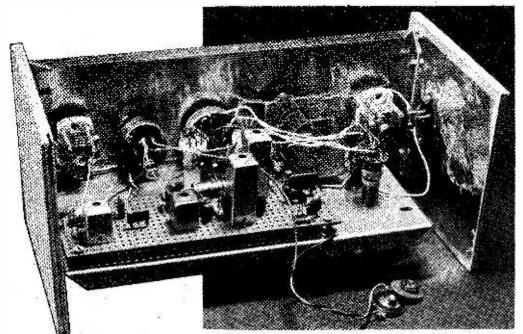
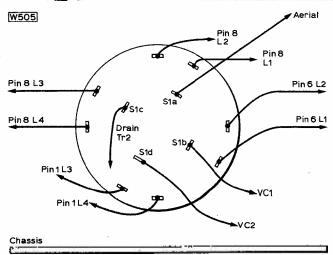


Fig. 4: above, shows the layout of the panel-counted controls and the passion of the mixer-IF bugget on the chassis. The president of the back of the president of the back of the packet also shows the packet of abundant of the case to stream the pois when the pre-selector is used with the receiver. Fig. 5 below helf, is the wiring between the band switch and the spur colls.

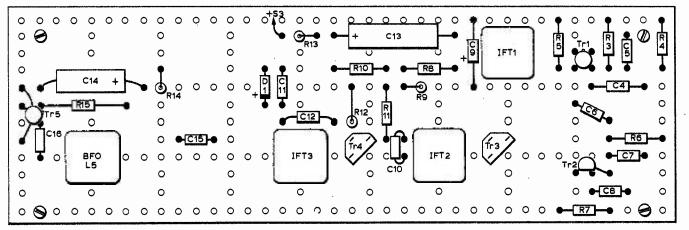


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operation. It is a good idea to place the audio board so that the lower cores of IFT1 and IFT2 can be reached through holes in the chassis, or to raise the IF board sufficiently to reach these cores. If the latter method is used, provide a return lead from the negative line to chassis. When the five cores have been peaked up for best results, they need no further adjustment. Alternatively, a signal generator, set to  $1.6 \mathrm{MHz}$ , may be used for precise alignment after the set is working.

#### **BFO Adjustment**

Initially tune in a steady AM signal precisely and set VC5 half closed, then switch on the BFO with S2. Set the core in L5 so that a strong audio heterodyne is heard and rises in pitch when VC5 is turned either way from the central, zero beat position. This oscillator is never in use for AM reception, only for CW



#### W506

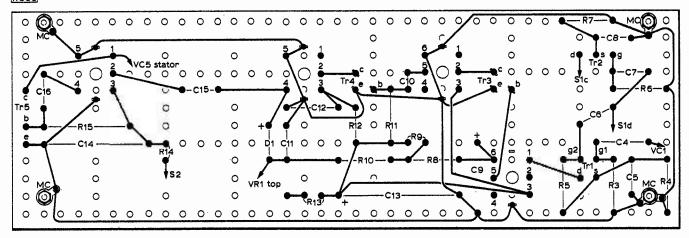
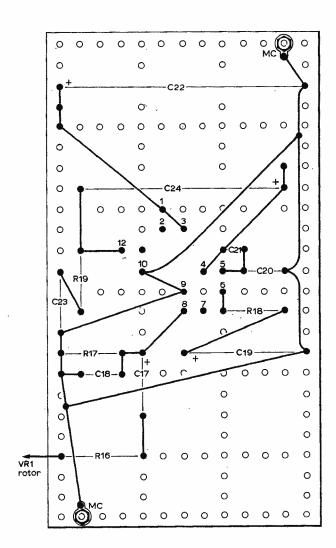
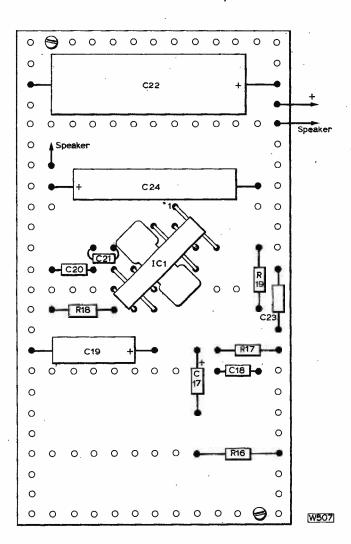


Fig. 6: Wiring and component location on the mixer-IF board is given here, from above and below. The drawings are not to scale

#### ★ components

Resistors	되어보는 그리다 사람들이 왕십다.	CASS NOT TOOK THE CASE OF THE
R1 1 2kΩ R8 47		- PAN <b>VC5</b> ( <b>15pE</b> ) (April 10 大きにが加づける。 日刊におり、2000年
R2 2 7kΩ R9 120		TC1 50pF compression trimmer
R3 120kΩ R10 27		Note:-1 nanofarad (1nF) = 1000pF
R4 2·7kΩ R11 330	Ok $\Omega$ R18 47 $\Omega$	
R5 100kΩ R12 390		Semiconductors
R6 1MΩ R13 890	OΩ VR1 10kΩ log pot	Tr1 40678 Tr5 BC107
R7 2·7kΩ R14 5·6	6kΩ with switch \$3	Tr2 MPF102 D1 OA90
All 1 or 1W 5%		Tra BF194 IC1 TBA800
Wit E OI 244 0 No		Tr4 BF194
Capacitors		**************************************
	C13 220µF 10V	Inductors
C1 33pF see text	C14 100µF 10V	IFT1 Denco type:IFT18/1-6
The state of the s		IFT2 Denco type IFT18/1-6
C3 100nF	C15 10pF silver mica	IFT3 Denco type IFT17
C4 300pF silver mica	C16 120pF C17 2µF 10V	# L5 Denco type IFT17
C5 20nF		L1 Denco range 4, blue, valve type
C6 27pF silver mica	C18 3 3nF	L2 Denco range 3, blue, valve type
C7 150pF ,,	C19 1000F 10V	
C8 20nF	C20 2·2nF de gassas de la company	L3 Denco range 4, white, valve type
C9 4aF 4V	C21 270pF silver mica	L4 Denco range 3, white, valve type
C10-20nF	C22 1000µF 10V	
C11 20nF	C23 100nF	Miscellaneous
C12 100nF	C24 470µF 10V	S1, 4-pole 2-way rotary switch. S2, on-off slide switch.
		S3, on-off switch, part of VR1. Perforated board 0-15in.
CP1 1nF silver mica	CP2 330pF silver mica	matrix 5‡ x 1f and 3‡ x 2in. Knobs. Flanged members for
	motion (Jackson type O2)	panel and chassis (Home Radio). Battery PP6 9V with
VC3 50pF variable (Jack	son type 804)	terminal connectors. Material for cabinet.
	·	





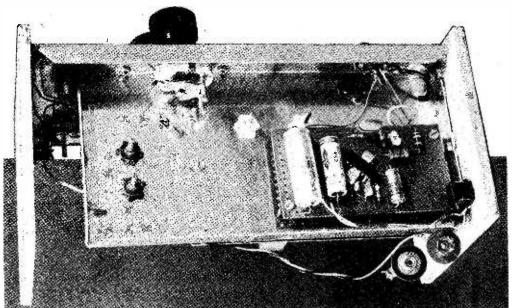


Fig. 1: above, is the piece of Veroboard. It 15 in. matrix, carrying the IC audio amplifier stage. Again, Ibls is not to scale. This board is mounted underneath the chassis as can be seen in the photograph, left. The variable capacitor, top left, is the Microtune of bandspread tuning control, VC4.

or SSB. For CW, set VC5 to produce the wanted audio pitch. With SSB, more critical adjustment of VC5 is necessary, to obtain the best resolution of speech. Remember that lower sideband is used below 10MHz and upper sideband above.

Due to the way in which D1 operates, satisfactory resolution on SSB makes it necessary that the signal

at the diode should not be too strong. When the RF stage is built, a control on this allows gain to be reduced. However, when the receiver is operated without this stage other means of reducing the input from strong SSB signals will have to be employed. Trouble from excessively strong signals at D1 will be unlikely with weak stations and a short aerial, but

will certainly occur with a better aerial, and with strong transmissions. A 50pF or similar variable capacitor in the aerial lead will then be worth while, or a small fixed capacitor as in Fig. 1.

#### Aerial/Oscillator Tuning

Trimmer TC1 is screwed down about half way so that when the coil cores are adjusted, signals peak up with VC3 about one-third to two-thirds closed. Band coverage is adjustable between quite wide limits and depends largely on the positions of the cores. For approximately  $1.7 \mathrm{MHz}$  to  $15 \mathrm{MHz}$ , set them so that about  $^{1}2$ in of the 6BA screwed rod projects. Subsequently, adjust the aerial coil cores, one for each band, so that VC3 allows peaking up of signals throughout each range.

Should alignment here be somewhat in error, it will be necessary to re-adjust VC3 for best reception when tuning. This will not cause any lack of efficiency, provided VC3 is neither fully open nor fully closed. Adjust the aerial coil cores near the low frequency end of the bands to minimise this. The bands covered are approximately 1.7 to 5.5MHz and 5.5 to 15MHz.

#### **Operating Notes**

Space is left inside the receiver for a PP6 9V battery. Any wire aerial, including an indoor wire, may be used. Long wire aerials should be coupled with a trimmer or other capacitor, as described, to help reduce the second channel interference which will otherwise arise. (This is greatly reduced by the RF preselector mentioned.) An earth connection will sometimes increase the volume of weak signals, especially at the lower frequencies. It is possible to arrange the coverage to tune to 1.5MHz if required, but whistles must be expected around 1.6MHz (the receiver IF).

A fairly large speaker in a suitable cabinet is best. High resistance or high impedance headsets can be used, provided a resistor is connected across them, about  $16\Omega$ . The combination is much less sensitive than when using  $16\Omega$  phones, but this is not too important because of the high audio output available.

Note that C1 cannot be left in circuit when using the preselector.

#### Tuning dial

The main tuning dial cursor is made from a small piece of thin clear plastic sheet. A hairline is scribed length-wise on the back. A hole is drilled to take a <sup>1</sup>4in. internal diameter grommet which is pushed on to the spindle of the tuning capacitor before the knob is fitted. The tuning scale is also clear plastic sheet, backed with white card.

#### Cabinet

The cabinet can be made from universal chassis members, or some parts may be of wood. Using 6mm ply, the sides should be  $4^{1}_{2}$  x  $4^{1}_{2}$ in each, and the top and back each 9 x 4in. The side near L1/L2 is lined with an aluminium plate or kitchen foil 4 x 4in to avoid coupling with the preselector coils when this unit is adjacent to the receiver.

Part 2 next month will give full details of the RF preselector for use with this receiver.

R.A.E.—continued from page 336

#### Squares and square roots

If you wish to square an index quantity it is necessary to double the index figure. For example:

 $10^3$  squared =  $10^6$  and  $10^{-6}$  squared =  $10^{-12}$ 

If you wish to find the square root of an index quantity it is necessary to half the index figure. For example:

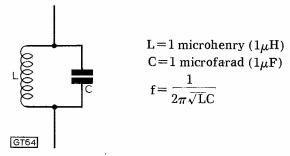
$$\sqrt{10^6} = 10^3$$
 and  $\sqrt{10^{-6}} = 10^{-3}$ 

If the index quantity contains a number, for example:

 $\sqrt{81\times10^6}$  this may be represented by  $\sqrt{A\times B}$ . First, square root part A, second, halve the index figure of B, thus:  $\sqrt{81\times10^6}$  =  $9\times\sqrt{10^6}$  =  $9\times10^3$  = 9,000.

#### Indices in calculations

Let's look at a typical problem that might arise in the examination, such as finding the resonant frequency of a capacitor—inductor combination,



If this equation is written out in figures, it is rather large and cumbersome, as shown below:

$$f = \frac{1}{2 \times 3 \cdot 142 \times \sqrt{0.000,001 \times 0.000,001}}$$

By using indices it is possible to make the whole calculation much simpler. So now re-write in index form:

$$f = \frac{1}{6 \cdot 284 \times \sqrt{1 \times 10^{-6} \times 1 \times 10^{-6}}}$$

add the indices:

$$f = \frac{1}{6 \cdot 284 \times \sqrt{1 \times 10^{-12}}}$$

square root by halving the indices:

$$f = \frac{1}{6 \cdot 284 \times 10^{-6}}$$

rearrange:

$$f = \frac{1 \times 10^6}{6 \cdot 284} = \frac{1}{6 \cdot 284} \times 10^6$$

 $f = 0.16 \times 10^6 \text{ Hertz}$ 

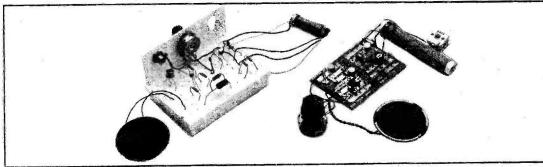
 $f = 160 \times 10^3 \text{ Hertz}$ 

f = 160 kilohertz (kHz)

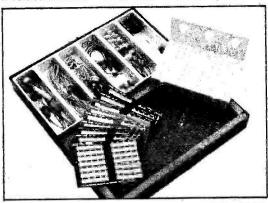
The values of L and C used in this example are not typical for this frequency but were chosen to demonstrate the use of indices in calculations.

Next month a start is made on Electricity and Magnetism, reviewing the characteristics of common components such as resistors, capacitors, transformers and magnets.

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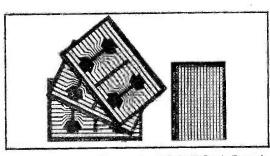


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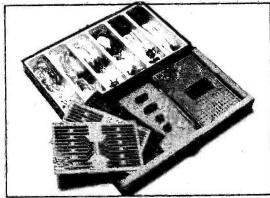
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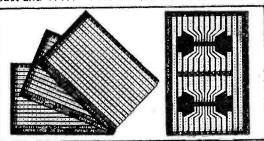
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18	4	2	4.03	0.96	3	2.0	5 · 27	0.96
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108	6 8	ă.	6.98	1.14	21	4-0	7.44	1-14
72	10	5	7 67	1.14	51	5.0	8 37	1.32
116	12	6	8 . 99	1 · 32	117	6.0	9.92	1 · 45
17	16	8	10-39	1.32	88	8.0	11 73	1.64
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226	60	30	26 82	OÁ		60 VOI	LT RANGE	
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	50 VOL1	RANGE	
Prim 22	20/240V Sec	0-19-25-33-4	10-50V
Voltage	s available	6, 7, 8, 10, 1	4, 15, 17,
19, 25, 3	31, 33, 40, 50	V or 25-0-25	V
Ref	Amps	£	P&P
102	0.5	3 · 41	0.78
103	1.0	4 · 57	0.96
104	2.0	6 98	1.14
105	3.0	8 45	1.32
106	4.0	10.70	1 · 50
107	6.0	14-62	1 · 64
118	6.0	17.05	2.08
119	10.0	21 - 70	OA

	S ISOLATIN		
P	RIM 120/240 S	EC 120/24	
Ref	VA (Watts)	£	P&P
07	20	4 - 40	0.79
149	60	6 - 20	0.96
150	100	7.13	1-14
151	200	11 16	1.50
152	250	12.79	1.84
153	350	16 28	2.15
154	500	19 15	2 · 15
155	750	29 - 06	OA
156	1000	37 - 20	OA
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159	3000	79 - 05	OA

	ISC	LTAGE MAI	
	Sec 100/1:	20V or 400/440 20V or 200/240	V
Va	Ref	£	P&P
60	243	5 · 89	1.32
350	247	14-11	1.84
1000	250	35-65	OA
2000	252	54 25	ÖA

Prim		LT RANGE Sec 0-12-15-20	0-94-30V
Voltag	es availabl	e 3. 4. 5. 6. 8	. 9. 10. 12
15, 18, Ref	20, 24, 30V Amps	15-0-15√, 12 £	-0-12 P&P
112	0.5	2.64	0.76
79	1.0	3.57	0.96
3	2.0	5 · 27	0.96
20	3.0	6 · 20	1.14
21	4-0	7.44	1.14
51	5.0	8 37	1 · 32
117	6.0	9.92	1 · 45
88	8.0	11 73	1.64
89	10.0	13.33	1 · 84

Prim		LT RANGE	0-48-60V
		e 6, 8, 10, 12,	
		60V or 24	
0-0-30	v		
Ref	Amps	£	P&P
124	0.5	3.88	0.96
126	1.0	5 · 58	0.96
127	2.0	7.60	1 · 14
125	3.0	10.54	1.32
123	4.0	12 · 23	1 · 84
40	5.0	13-95	1 · 64
120	6.0	15 66	1.84
121	8 0	20 15	OA
122	10.0	24.03	OA

1	2.0		27.13		UA
AU'	TO T	ANS	FOR	MERS	
VA	(Watts	) Volt	s	£	P&F
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75	0-115-	210-2	40	3.59	0.9
150	0-115-	200-2	20-240	5.35	0.9
300	0-115-	200-2	20-240	7.75	1.14
500	0-115-	200-2	20-240	10.00	1 . 64
1000	0-115-	200-2	20-240	18.76	2.08
1500	0-115-	200-2	20-240	23 - 36	OA
2000	0-115-	200-2	20-240	34 82	OA
3000	0-115-	200-2	20-240	48.00	OA
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207	500, 500	0-8-9-, 0-8-9	2.50	0.71
208	1A, 1A	0-8-9, 0-8-9	3 - 53	0.78
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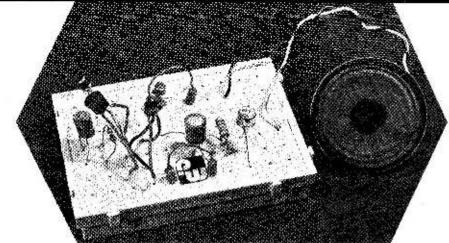
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# S-Pecaelogy



DAVID GIBSON

10



Record Player Amplifier'

This month's project is a simple record player amplifier which is suitable for use with a ceramic cartridge. Although the output could not be fairly described as hi-fi, it does give useful volume which is acceptable in terms of quality.

The prototype used a small 58mm diameter  $80\Omega$  loudspeaker but a larger diameter unit would undoubtedly improve the apparent volume and quality, particularly in the base notes. Any speaker with an impedance from  $60\Omega$  to  $80\Omega$  will be satisfactory.

#### Circuit description

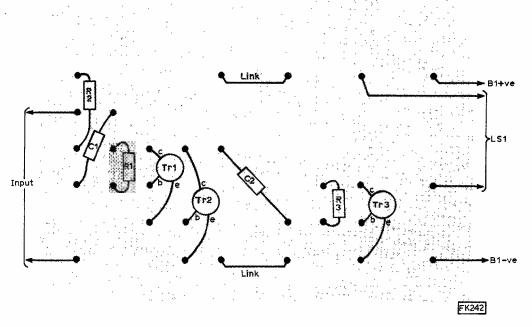
The circuit, Fig. 1, is basically a two-stage amplifier. Transistors Tr1 and Tr2 connected in a Darlington pair configuration form the first stage. This useful arrangement has two main advantages. The first is

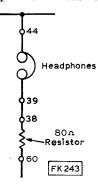
that a great deal of amplification is possible. The total current gain from such a circuit is found by multiplying the gain of Tr1 by that of Tr2. In practice it is difficult to calculate exactly what the gain will be because individual transistors have quite a spread in their gain characteristics.

Secondly, the Darlington stage also keeps the input currents very small indeed. In terms of power consumption the two Darlington connected transistors drew less than one fortieth of the total current of the whole amplifier.

Resistor R2 forms an audio load for the Darlington pair while base bias and DC stabilisation is provided by negative feedback resistor R1.

The first stage is capacitor coupled to the output stage, Tr3 via capacitor C2. Again, in the interests of simplicity, a lone resistor R3 provides bias and





The left-hand drawing is a full-size S-DeC layout showing component locations. The smaller drawing (above) shows how a series resistor of about 80Ω must be used when operating the amplifier with low impedance headphones.

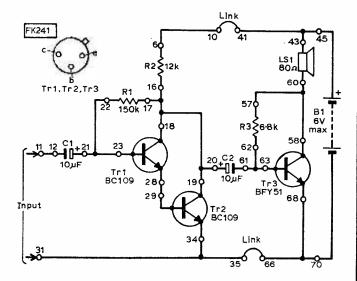


Fig. 1: Circuit diagram of amplifier.

stabilisation. There are two important points to note. First, do not use very low impedance speakers in this circuit, i.e. 3, 4, 8 or  $16\Omega$  units. Secondly, note that both capacitors are electrolytics and must be connected with correct polarity. The positive lead of C1 must go to S-DeC hole 21 and the positive lead from C2 must be pushed into hole 20.

Remember the two shorting links which connect together holes 10/41 and holes 35/66 respectively. It is all too easy to forget these since they are not

actual components.

This little amplifier works well from any voltage between 1.5V and 6V although sensitivity and output are lower at the lower voltages. At 1.5V the circuit draws approximately 6.5mA; at 3V it is 16mA, at 4.5V it is 25mA and at 6V, 34mA. With 6V applied, the current taken by Tr1 and Tr2 was found to be well under 1mA.

The lead from the record player pick-up should be of the screened variety. The centre conductor should be connected to hole 11 while the screening wire mesh goes to hole 31.

### you will need . . .

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The amplifier provides an excellent and inexpensive general purpose unit and to this end is probably well worth transfering to a piece of Blobboard to form

a permanent piece of equipment.

A crystal microphone connected to the input (holes 11 and 31) worked well; the unit would make a very useful intercom amplifier by using another, similar speaker as a microphone and connecting this to the input. The amplifier could also be used as the audio section of a radio receiver—even a crystal set—and headphones may be connected if required. Most types offered for hi-fi and record monitoring purposes are of the low impedance kind and it would be wise to insert an  $80\Omega$  resistor in series with the phones if this application is envisaged.



#### ONE-CHIP TOUCH TUNING

Mechanical station selection using a combined potentiometer/switch assembly is the most common method employed in sets equipped with varicap tuners. Our constructional project next month describes an exceptionally easy method of going "all electronic", based on a Plessey touch-tuning i.c. Connection to the receiver is via the existing leads used for the mechanical selector. The unit is powered from the 33V tuning supply and uses l.e.d.s for channel indication.

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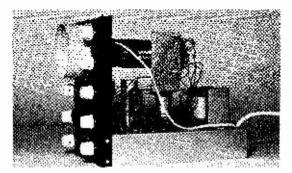
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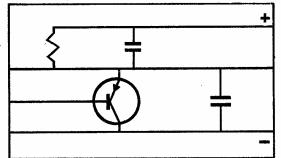
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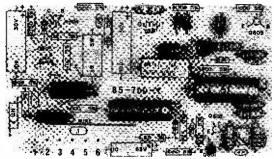
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Hundreds of new components—pots, switches, resistors, capacitors, PC Boards with semiconductors, loads of odds and ends. Amazing value at only £3·45.

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Large quantity always available; 3lb asstd. £1-90; 7lbs £3-90; 56lb £17. Pack with about 500 components inc. at least 50 transistors £1-10; Pack with 12 High quality panels, incl. IC's power transistors, multiturn trimpots, hundreds of small signal transistors, resistors, zeners, capacitors, etc. Only £2-95.

7400 IC pack—20 74 series IC's on panels £1.25.

#### PC ETCHING KIT MKIII

Now contains 200 sq. ins. copper clad board. 1lb Ferric Chloride, DALO etch-resist pen, abrasive cleaner, two miniature drill bits, etching dish and instructions. £4.25.

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Anhydrous technical quality in 11b double sealed packs. 11b £1·10; 31bs £2·30; 10lbs £5·80; 100lbs £39·00.

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Large quantity of offcuts, all usable pieces. 200 sq. ins. single sided, double sided or mixed £1.50.

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50V ceramic plate capacitors, 5% 10 of each value 22pF to 1000pF. Total 210 caps.

z.3·35.
CR25 carbon film \( \frac{1}{2}\) watt resistors, 5% 10 of each value 10 ohms to 1M, Total 610 resistors \( \frac{25}{26}\) 00. Electrolytics, wire ended, 25\t working 10 each of \( \cdot 1\), 2·2, 2, 4·7, 10 22, 47 and 100mfd.

22, 97 and 100mtd.
70 capacitors for £3.50. 400mW zeners 5% 10 each 3V to 30V, 250 in all. Only £15.30.
1 watt Metal Film Resistors, 5% 10 of each value 27R to 10M. E12 series. Total 680 resistors £12.00. Polystyrene capacitors, 160V 21% 10 of each value from 10pF to 10,000pF.

Total 370 capacitors £12:30.

#### COMPONENT PACKS

400 asstd, carbon resistors	£1 · 50
100 Wirewounds 2-15W	£1 · 50
200 Miniature resistors, 1, 1, and	}W
	£1 30
200 poly, mica, ceramic caps	£1 · 20
100 polyester, 01-2-2uF	£1 · 00
200 PC resistors	75p

TAPE DECKS

Data Tape Deck by D-Mac, Reel-to-reel, takes 3" spool. 2 heads, 12V motor. 3 position switch. 17 transistors, R's, C's 200V 6A SCR, min 6V relay, etc. in good condition but not new. Size 225 × 185 × 120mm high. Reduced to clear £4.

#### **TRANSFORMERS**

All have mains primaries, secondaries as follows: 6-0-6V 100m A 95p; 9-0-9V 75mA £1; 12-0-12V 50mA 95p; 12-0-12V 100mA £1·10; 6-3V 1½A £1·95; 6-0-6V 1½A £2·30; 9-0-9V 1A £2·55; 12-0-12V 1A £2·35; 12V 5A £3·85; 12V 8A £4·35; 20V 55mA 90p; 22V 100mA £1·09; 29V 50mA 30p. Multitapped types, 0-12-15-20-24-30V 1A £3·52V 100mA £1·09; 29V 50mA 30p; 100mA £3·52V 50mA £1·09; 20V 50mA 50p; 100mA £3·52V 50mA £3·52 tapped types, 0-12-15-20-24-30V 1A £3·50; 2A £5·00; 20-0-20V 2A £4·40; 24V 500mA £1·85; 20V 2}A £4·15; 14V 1A £2·00.

MULTIMETER MODEL LT 22

#### **AMAZING** VALUE !!



Probably one of the lowest priced meters featuring a mirror scale, overload protection and 20000 ohms per volt sensitivity. A modern styled meter with unsually clear scale markings. Ranges: DC Volts 0-0-5-2-5-10-50-250-1000 AC Volts 0-10-50-250-1000 DC Current 0-50uA-25mA-250mA. Resistance 0-50k-50k-50k-50k (Mid scale 300-3k-30k) db's 4 ranges 20db o +42dB. Size 130 × 90 × 42mm £12·70.

#### CALCULATOR KEYBOARDS CLEARANCE

Many different types, some as previously advertised at up to £1.80, some off Rapidman calcs. All types at one price £1.20 each.

THE NEW 1977 GREENWELD CATALOGUE WITH DISCOUNT YOUCHERS IS NOW AVAILABLE—PRICE 30p + 15p post.

#### ELECTRONIC CAR VOLTAGE REGULATOR

All parts for the project featured in this edition of PW available in kit form for only £4.00.

240V a.c. 60 r.p.m. High Torque: drive to 6mm shaft 20mm long. Size 70mm dia.  $\times$  55mm £2:20.

#### LED DIGIT DRIVER

ITT type 7105. 16PIN DIL package. Supplied with Data sheet. 8 for £1.

#### **VERO OFFCUTS**

Pack A, All 0·1" Pack B, All 0·15" Pack C. Mixed. Pack D, All 0.1" plain

Each pack contains 7 or 8 pieces with a total area of 100 sq. ins. Each pack is £1·40. A iso available by weight, 11b £3·45, 10lbs £28, (0·1" only).

We are also Vero wholesalers, trade price list on request from Bone Fide Companies.

#### VEROCASES

Plastic top and bottom, ally panels front

pe 1410 205 ×	140 × 40mm	£3·70
1411 205 ×	140 × 75mm	£4·17
1412 205 ×	140 × 110mm	£5·20
1237 154 ×	85 × 40mm	£2 83
1238 154 ×	85 × 60mm	£3·05
1939 154 ×	85 × 80mm	£3.7

#### **VERO PLASTIC BOXES**

Professional quality, two tone grey polystyrene with threaded inserts for mounting PC Boards.

Type 2518	-120 ×	65 ×	40mm	£2·42
2520	150 ×	80 ×	50mm	£2·68
2522	188 ×	110	60mm	£3·72



Sloping front version.

Type 2523 220 × 174 × 100/52mm £6.90 1798 171 × 121 × 75/37 5mm £4 65 1/98 / 1/ × 121 × 1/3/3/ 5mm 24\*09

Gen Purpose plastic potting box 71 × 49 × 24mm in black or white 40p Hand Controller box, shaped for ease of use in the hand, 94 × 61 × 23mm 64p.

#### **SOLDERING GUN**

Instant heat soldering gun, mains voltage, built in illumination, 5 spare tips. ONLY £4.00.

£4·00.

S-DEC £2·20; T-DEC £3·85. Insulation tape, 15mm wide × 5m long, 7 diff coloured reels £1; Switch banks, 4 PO 1000 type lever switches in vertical bank. 3 for £1·20. Miniature carbon film 5% resistors, all values from 1R to 10M 1½ p each, 100 of one value £1.

#### **5V 12A REGULATED** POWER SUPPLY

Brand new boxed fully stabilized PSU. Complete with instruction manual. Load regulation 0·15% Thermal & Electronic overload protection. Only £24·00.

Our retail shops at 21 Deptford Broadway, London, SE8 (01-692 2009) and 38 Lower Addiscombe Road, Croydon (01-688 2950) stock some of the advertised goods for personal callers only. Ring them for details.

All prices quoted include VAT and UK/BFPO postage. Most orders despatched on day of receipt. SAE with enquiries please. MINIMUM ORDER VALUE £1. Official Orders accepted from Schools, etc. (Minimum invoice charge £5). Export/Wholesale enquiries welcome. Wholesale list now available for bona-fide traders. Surplus components always wanted.





4 Red 1 inch high LEDS. 12 hour display with AM/PM indication Power failure indicated by flashing display. Precise accuracy from mains frequency. Beautiful Brayan Teak case or stylish Perspex (State first & second colour

NON ALARM:

Complete kit including Teak Case £11.50 incl. Module Kit excluding case £9.00 incl.

ALARM EXTRAS: Pulsed alarm tone. Tilt operated "Snooze" period. Automatic brightness control. Simple setting.

ALARM:

Complete kit including Teak case £14·50 incl. Module Kit excluding case £12·50 incl.

PERSPEX CASES:

50p LESS than Teak.

READY BUILT:

Extra £2:00 on complete clocks, Extra 50p on Modules.

TIMER FACILITY:

Count in seconds up to 9m 59sec. Extra 50p.



#### "ALPHA"

£11.00 incl.

Ready Built. Non Alarm. Perspex. 4 digits phosphorescent-green. \( \frac{1}{2} \) high 12 or 24 hour display. White, black, blue, red, green.

Discount on orders over 5 clocks or kits. Overseas: Add £2:50 for Airmail.



Send payment with order. SAE for details.

(PW 1) 6 GOWER RD, SOUTHVIEW HOUSE, ROYSTON, HERTFORDSHIRE Phone Royston (0763) 43695

# PRODUCTION LINES bill tull

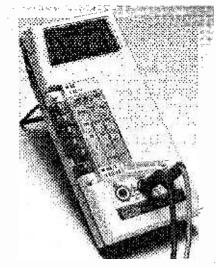
#### A free hand from Fluke

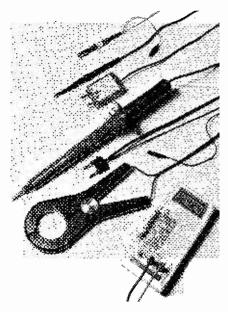
Possibly a little on the expensive side for a lot of our readers, but if you are in the market for a new multimeter, don't rush out and buy the first you see, think about the new Fluke model 8020A digital multimeter.

Incorporating a  $\frac{1}{2}$ in  $3\frac{1}{2}$  digit LCD, the various functions and range pushbutton selectors are so arranged as to allow one-handed operation. The latest CMOS/LSI circuitry is used throughout, and this together with an LCD, ensures that the small 9V alkaline cell lasts for up to 200 hours.

Apart from the usual features of a multimeter such as ten voltage ranges, six resistance ranges, and current measurement up to 2A, the 8020A incorporates three diode test functions and two conductance ranges. In the case of the former of these two functions, the meter applies sufficient voltage to turn on a semiconductor

junction, so that diodes and transistors can be tested for the correct forward bias voltage in situ. Individual paralleled resistors can also be checked without the need to disconnect them from the circuit.





The two conductance ranges enable measurement of resistance as high as 10,000M ohms, thus making possible the checking of HV dividers, leakage in capacitors, PCB's and cables.

Overload protection is also part of this new meter, with full protection against 300V DC continuous, or 6kV transients on any range. Priced at £99 and guaranteed for 2 years the 8020A is complete with test leads and instruction manual, although the versatility of this meter could be extended by purchasing a high voltage probe, RF probes, temperature probe and external battery eliminator.

ITT Instrument Services, Edinburgh Way, Harlow, Essex. Tel: 0279 29522.

#### IC test clip

Walmore Electronics have introduced a unique DIP test clip which is designed to provide positive contact to eliminate the possibility of accidental disconnection from a device. Called the DIP/LOC it has contacts which provide a wiping action with the DIP leads and end-contacts hook under



the DIP's end-leads in a lock-on action.

14 and 16 pin versions are available, while cable connections can be made quickly and easily with barrier strips between contacts to prevent lead shorting.

Walmore Electronics Ltd., 11-15 Betterton Street, Drury Lane, London.

#### Cassette centre

Not everyone wants a mini-recording studio in their sitting room as would appear from the complexity and price of many of the commercially available cassette decks. Plustronics, who have recently acquired the Aiko brand, are now marketing the Aiko ATP-711 stereo cassette recorder which is complete with twin speakers and microphones, and would suit the person who requires a good quality recording system, and yet is not too expensive.

Power output into the 8 ohm speak-

ers is claimed to be 5W per channel while frequency response is 50 to 12,000Hz. Controls are of the slider type, with recording level monitored



by twin VU meters. Priced at around £117 the Aiko ATP-711 can be obtained from the usual Plustronics stockists.

Plustronics Ltd., Hempstalls Lane, Newcastle, Staffs. Tel: 0782 615131.

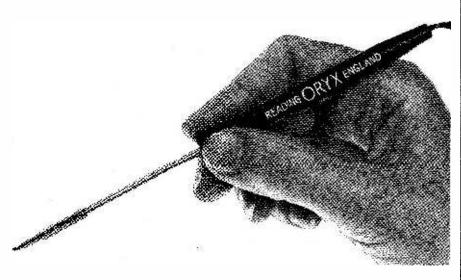
#### Flyweight iron

Components seem to be getting smaller and smaller these days, and this coupled with the fact that many of todays IC's are extremely heat and voltage conscious makes the job of soldering a hazardous affair. Now, thought to be the smallest iron in the world, comes the Oryx model 9 low voltage iron that weighs in at only 6 grammes.

Manufactured by Greenwood Elec-

tronics, the element is claimed to give an improved and longer working life, and is available in either 6V, 12V or 24V models. The maximum tip temperature reaches 325°C while power consumption is about 8.3W. The complete iron is only 160mm long and costs £3.02 from the manufacturers or from the usual retail stockists.

Greenwood Electronics, Portmann Road, Reading, Berkshire, RG3 1NE. Tel: Reading 595844.



#### Multi-pin resistor

A new style of resistor recently to enter the UK market, has been launched by National Semiconductors. The resistors comprise 0.1in pitch SIP resistor networks available in 6, 8, and 10pin epoxy coated packages, with two styles being offered for each size package. In styles 605, 807 and 109, pin 1 is common to all resistors in the package, whereas in styles 603, 804 and 105 each individual resistor is brought out to a separate pin.

Line terminating resistor networks are available in 8 and 10pin styles with resistor values R1/R2 being 180/390 ohms or 220/330ohms respectively. All styles are rated at 0.125W, and have a tolerance of  $\pm 2\%$ .

Any further information from National Semiconductors Ltd., Stamford House, Stamford New Road, Altringham, Cheshire. Tel: 061-928 3417.



#### A good case

Recently introduced by Boss Industrial Mouldings is a new range of aluminium containers, notated the BIM7000 range with a choice of either 15° or 30° sloping front panels. They comprise two sections each and are supplied in blue/off white, green/sand and gold/satin black colour combinations for base and sloping panels respectively.

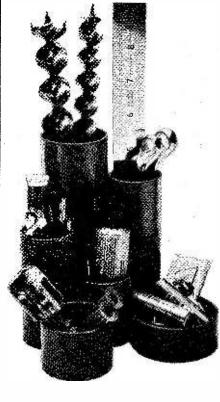
16 sizes are available, all of which have ventilation slots in both the base and rear vertical panels.

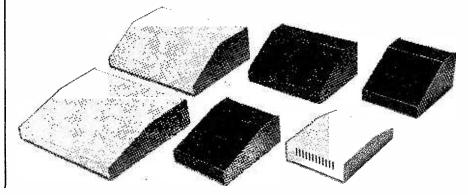
Boss Industrial Mouldings Ltd., Higgs Industrial Estate, 2 Herne Hill Road, London, SE24. Tel: 01-737 2383.

#### Tidy tubs

Take a look in one of your drawers in the kitchen, the chances are that if not all of them, at least one of them will be full to overflowing with bits and pieces. pens, pencils, clips and other things which may "come in handy" at some future date. Any attempt to tidy up results in the mess being transfered to another draw, but it's just possible that a new device produced by Platignum (the pen people) will solve this age old problem. Called Tidy-Tubs, it comprises six plastic tubs of varying dimensions joined together to form a "hold-all" for all those small bits that lie in the bottom of draws.

Tidy-Tubs come in six colours, cost £1.50 each, and should be available from all good department stores and stationers.

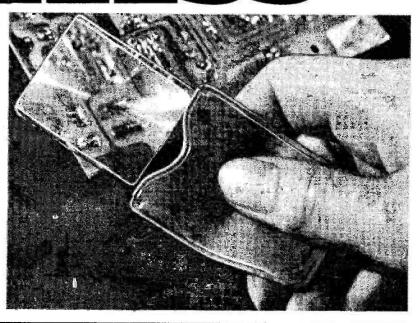




NEXT MONTH IN...

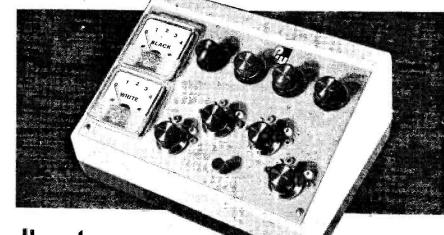
# ON SALE SEPT. END.

Now! You can avoid disappointing results with that new project by checking for dry joints and solder bridges on your printed circuit boards with our FREE Pocket Magnifier. Inspect your handiwork in detail. There's a dozen other uses for this handy gadget.



## 

Losing a few friends through boredom seems to be a hazard of some current "logical" games. October PW features SOLO SUPERMIND, which not only makes the "codemaker" redundant, but permits you to set and break your own codes. Simple circuitry, using readily available transistors disposes of the need for expensive "chips". It's effective, engaging, and cheap to build.



Audio Level Indicator Inexpensive Sine~Square Wave Generator The Jubilee is an electronic chord organ with automatic vamping coupled to a three instrument rhythm section. It incorporates a number of features normally found only on the more expensive commercial instruments such as variable sustain length, percussion and other effects. Although automatic organs of this type have been available on the commercial market for a few years they have previously been put out of court for the home constructor because of their inherent complexity and high cost. Nevertheless we have overcome some of the problems with a novel design; simplicity of construction has been obtained by using three specially designed organ integrated circuits—developed for the professional manufacturer General Instrument Microelectronics. These revolutionary integrated circuits are now available to the amateur.

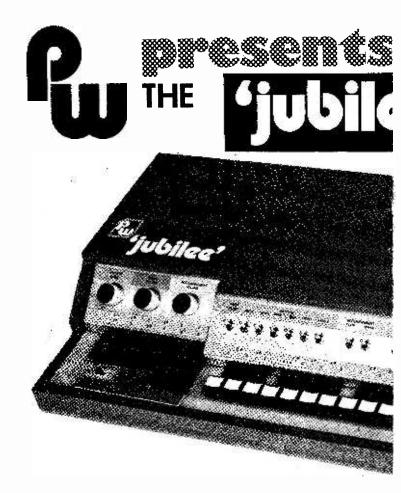
#### Vamp playing

The author, like many readers, is fascinated by electronic organs and the sounds that they produce but, unfortunately, is not sufficiently dextrous to get the best out of a conventional instrument. For this reason every effort has been made to design an instrument that will produce a wide variety of sounds and which can be played by the absolute novice. In practice the melody of a tune is picked out by one finger of the right hand and a four or five note chord (Major, Minor, Major 7th or Minor 7th in any key) is selected by one or two fingers of the left hand (the latter depends on the constructional method adopted as will be seen from reading on). The left hand chords can be played in a static form or can be vamped with an alternating bass note by selecting a suitable rhythm on the tab switches.

A vast range of music is available that enables the person with no previous experience to learn to play in a few minutes. Perhaps the best examples are those written with the Lowrey and Farfissa automatic organs in mind; these scores are sometimes referred to as 'Minute Music' or 'Speed Music'. Apart from these specially written music sheets it is easy to play any music that contains guitar chord notation. The Jubilee will capture the interest of any child and, although it is no substitute for proper music lessons, it provides an incentive to learn to play a more conventional instrument in due course. For those of us with somewhat longer teeth it is a superb fun instrument providing hours of relaxation which can be appreciated by other members of the household!

#### **Keyboards**

Again, like many readers, the author dislikes messy constructional projects which involve lots of discrete point-to-point wiring. This is usually typical of organs and, apart from leaving plenty of scope for error, it turns an interesting project into a bit of a chore. For this reason we have tried to keep construction to the assembly of components on a single printed circuit board and this includes the novel keyboard arrangement suggested for the *Mark I* version. In this example we have used calculator keys instead of a normal keyboard which would need a wiring loom. Calculator keys can be used quite satisfactorily because generally speaking only one has to be depressed at a time and slick fingering is not necessary because of the built-in sustain facility.



### M.J. HUGHES, MA, C. Eng., \* specifications

Three octaves for melody line with three principal tone colours plus any combination of these.

Variable length sustain.

Variable depth Vibrato.

Percussion effect.

Repeat percussion effect (for banjo and mandolin sounds by judicious use of length sustain can also simulate xylophone, glockenspiel etc.).

Variable repeat rate.

Two octaves of accompaniment.

Automatic chording giving all major, minor, and seventh chords.

Two principal tone colours plus their combination.

Percussion effect gives piano, guitar or banjo accompaniment.

Automatic vamping with alternating and walking bass. Rhythm section comprising bass drum, snare drum, and cymbals.

Six selectable rhythms—synchronised with vamping accompaniment.

Rhythms may be mixed.

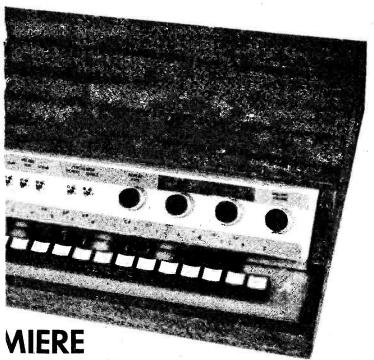
Variable rhythm speed.

Independent volume controls for melody, accompaniment and drums.

Internal 1W amplifier drives loudspeaker direct or phono output.

Single board construction—no wiring loom needed for keyboard.

## ELECTRONIC ORGAN Part 1



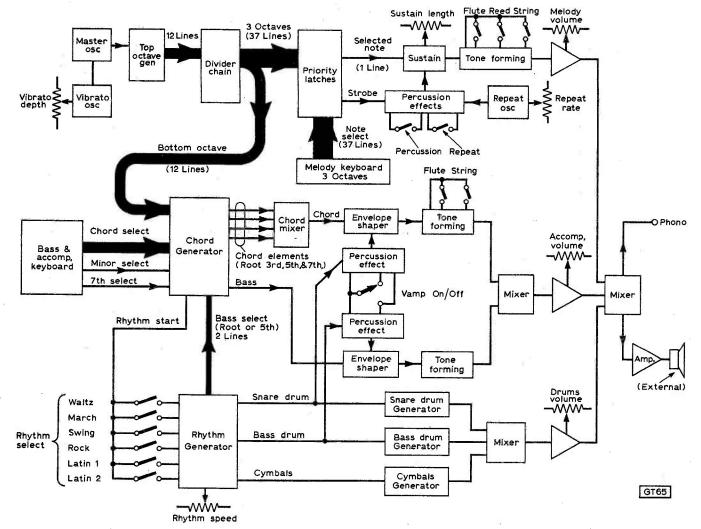
#### \* collated components list

Resistors .	
1 off 5-6Ω 4W	34 off 1040
3 off 2200	"t off text
1 off 3300	11 off 22kΩ
1 off 470Ω	1 off 33kΩ
3 off 6800	1 off <b>6</b> 8kΩ
2 off JkΩ	10 off 100kΩ
2 off 1:2k()	<b>2</b> off 470kΩ
1 off 2-2kD	3 off 1MD
4 off \$ 3kQ	roff 5 3MQ
11 off-4 7kΩ	
All 10% IW except when	stated
Capacitors	
5 off inF Polystymne	1 off the
3 off 33nF Polysytrene	2 off 2-2µF 15V Electrolytic
2 off 10nF	1 off 10µF 25v Electrolytic
1 off. 15nF	1 off 47µP.25V. Hectrolytic
3 off 22nF	1 off 100μF 25V Electrolytic
1 off 38nF	, 1 off 1000μF 25V Electrolytic
10 off 100nF	1 of 2200µF 25V Electrolytic
5 off usin F	2 of 4700pF 25V Electrolytic
10 off 470nF	
All polyester 10% (voltag	e preater than 25V) except where
statud	
Transistors	Diodes
1 off BFY71	6 off 1N914 (or 1N4148)
8 of 5C148	1 off 3V 400mW zener
	2 off DIL 1 Amp bridge type VM18 (Marshalls)

Preset patentiometers	Spindle potentiometers
1 off 5kΩ	2 of 5Ω lin
3 off 10kQ	3 off 10 kΩ log* .
3 off 100kΩ	1 off 100 kΩ lin
5 off 500kΩ	1 off 1 MΩ lin
All sub miniature horizontal	* One of these should have
	ganged mains switch for
	conventional Reyboard
	version.
Integrated circuits	2.4
7 off 741 1 off 4	Y-1-0212 (Top octave penera-
in the second second	
	Y-1-5050 (7 stage divider)
1 off 4011 3 off A	Y-1-1813 (Priority fatches)
Toff LM380 Toff A	Y-5-1317A (Chord generator)
1 off 7812	Y-5-1315 (Rhythm generator)
	y, Pattern 004 Jubilee' for the
Jatte	
NOTE: All devices prefix	ed 'AY are by General Instru-
ments. These devices a	re availa le ir m Technomatic
Ltd., Watford Electronic	s and other advertisers in the
magazine.	
Miscellaneous	
	r (Double 12V 1A secondary)—
Douglas type MT	ay sub-min toggle switches
**1 off double note chan	ge over sub-min toggie switch
*51 off AB Metals low o	rofile calculator buttons (0.6"
square) + available	from WKF Electronics.
Whitwell, Near W	forksop, Notts.
1 off 1A mains fuse an	d panel mounting helder
*1 off mains slide switch	h i i i i i i i i i i i i i i i i i i i
🥭 , I off Indicator lamp and	d holder (optional)
	erd (Reader's PCB Service)
DIL sockets or Soldercor	is for IC mounting
Boàrd pins (quantity depr	ends on version being built)
Standard lack socket 1" (	lor (oudspeaker output)
5-pin 180° DIN socket (for	
it the conventional ke	shoard version is built the 14
taggie switches can be	raplaced by organ rocker tabs
and the calculator keys	by a four octave keyboard with acts having common busbars.
strigte pole trake com	ch is replaced by a panged
evitch as molecus well	me polentiometer. Extra com-
non-rite regulated for th	e conventional version are:
ctoren lark ancket	(for link to foot controlled
Minor and Seventh	witches)
Pair of foot operated	push-to-make switches
	es are able to supply suitable
foot switches. They also h	rave limited stocks of 4 octave
keyboards having the lowe	est octave in reversed colours
(white accidentals). They c	an supply gold wire for making
single pole contacts or,	alternatively, low cost contact
blocks can be obtained fr	om Harmonics (Bromley) Ltd.
Oakley Grange, Plaistow	Road, Kirdford, Billingshurst,
West Sussex. The latter re	guiré palladium busbars which 🧎
Harmonics can also supply	

Some people might feel that they would prefer a conventional keyboard—certainly the sound of the instrument warrants the extra cost and trouble involved—and if so they can use the same printed circuit board with no changes except for the omission of the calculator keys. Flying leads are taken from each key position to the respective single-pole single-way contacts under each note of the keyboard.

In this part we shall describe the overall workings of the instrument and pay special attention to the large scale integrated circuits used. Constructional details will start next month and in order to allow time for ordering and delivery we are publishing a collated components list immediately. This might enable constructors to obtain components on favourable terms if they are purchased as a package deal.



#### System organisation

Constructors who are more inclined to build a conventional organ, or who have already built our previously published *Easibuild Organ*, might find this single board useful in providing an additional keyboard for solo or special effects. In this application some economy can be effected because the tone generator circuitry would already exist.

A block diagram of the complete organ is shown in Fig. 1. A master oscillator, frequency modulated at variable depth to give vibrato, drives a conventional top octave generator. Each output of this is fed to a two stage divider except for 'C' which is followed by three stages of division to provide a total of three complete octaves plus 'Bottom C'. These notes are used to drive the priority latches for the melody keyboard and the bottom of the three octaves is fed to the chord generator.

The priority latches are contained in three integrated circuits and in simple terms they are flip-flops which control a gate for each note. Each key of the melody board has its own flip flop/gate arrangement with the associated note fed from the tone generator to that gate. When the key is pressed the flip-flop opens the gate and allows the note to appear at the single line output. At the same time any other flip-flop is reset. This means that only one note at a time can be selected by the melody keyboard—it is monophonic. This, at first, sounds as though the scope of

Fig. 1: Block diagram of the complete organ.

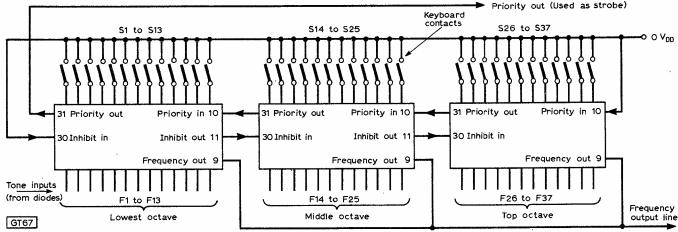
the organ is immediately limited. On the contrary. If the key just selected is released the flip-flop remains in its set state and the note remains active at the output and enables the envelope generator (built into the sustain circuitry) to hold the note over an adjustable period. This allows one note to die away slowly until a second key is pressed.

In order to synchronise the Sustain circuitry with the note selection operation we have to tell it whenever a new key has been depressed. This is easily done by making use of the internally generated 'Priority Output' signal for the priority latches. This signal in conjunction with the 'Inhibit Output' line is used within the priority latches to carry out the reset operation already referred to.

The three priority latch integrated circuits are coupled together in such a way that if two keys are pressed at the same time only the higher note will be selected. A pin connection diagram for this integrated circuit and an illustration to show how the three are coupled together in this application are shown in Fig. 2.

#### Sustain

The synchronising signal from the priority latches is called a 'Strobe'. This is fed to the percussion effects stage which changes the nature of the signal. Normally the signal is high for as long as a key is pressed—thus allowing a continuous tone through the



sustain circuit. The percussion effect, when switched in, allows only a short pulse from the strobe line and this occurs on the instant of pressing the key. The effect of this is that the note is generated for a very short period (even though the key is held pressed) and dies away according to the setting of the sustain length. Depending on the setting of the tone forming circuit the sound so produced could be like a drum stick hitting a wooden block if there is zero sustain length or like a bell being struck if there is a long sustain.

Another feature of the percussion circuit is the ability to repeat the pulse, just referred to, at an adjustable rate for as long as the key is depressed. By careful setting of the various controls one can produce sounds ranging from a xylophone to a banjo being strummed. With the repeat rate set close to the vibrato frequency very weird sounds can be produced!

#### Voicing

The shaped signals from the sustain stage are then fed to the tone forming circuits. We used the absolute minimum of circuitry here to keep construction simple but have managed to provide three basic tone colours. In electronic terms these are a heavily filtered square wave which gets close to sinusoidal, a straight square wave, and a spike waveform. These closely resemble the tone colours usually referred to as Flute, Reed and String respectively. The apparently small choice of tone colours is deceiving. By modifying their envelopes, they provide a wide range of sounds

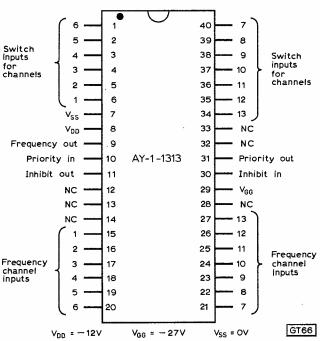
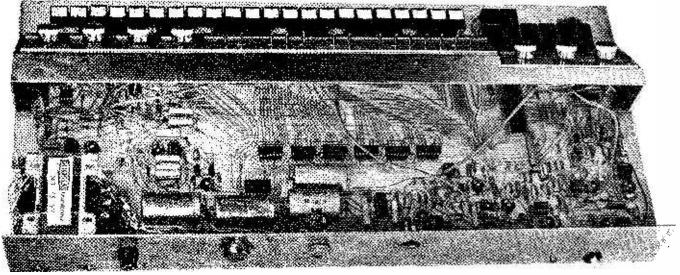


Fig. 2 (above and top): Pinout and simplified interconnection diagram for the three priority latch IC's AY-1-1313. Among other functions, these circuits ensure that, if two notes are pressed at the same time, only the higher one will sound.

The picture (below) shows a prototype organ and has been included solely to give readers some idea of the work entailed in construction. The IC's referred to above are hidden from view under the front panel hood.



Practical Wireless, September 1977

ranging from ecclesiastical to modern plus the instrumental effects, some of which have already been mentioned. The signal is then fed to a pre-amplifier where its volume is independently controlled prior to joining with other signals at the power amplifier.

#### Chord and rhythm generator

The more complex part of the instrument is the accompaniment section which is built around General Instrument's chord and rhythm generator ICs.

As the chord generator concept is relatively new to the amateur it is worth spending a few paragraphs describing how this chip works. Its pinning diagram is shown in Fig. 3. The generator can be used in two configurations. One requires 36 separate keys to select the major minor and seventh chords for a complete octave. The alternative configuration re-

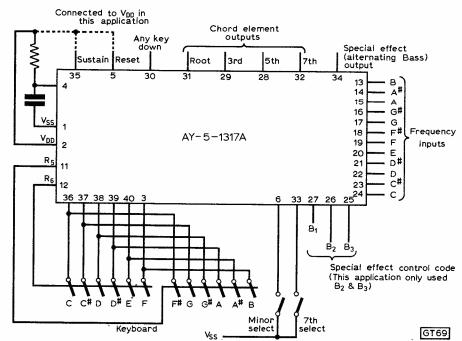
quires 12 contacts to select one of the major chords and utilises two extra switches to convert the selected chord to minor and seventh respectively. The latter configuration is used to economise on keyboard facilities. This means that not all the input pins of the package are utilised. The schematic stage shown in Fig. 4 might be of more help in describing how the device functions in our configuration.

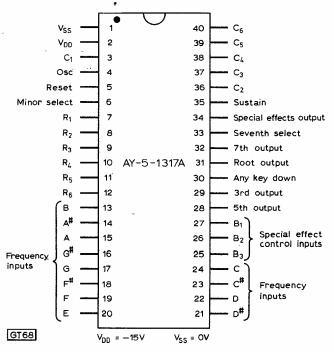
The twelve notes of the bottom octave from the tone generators are connected directly to input pins 13 to 24 and a capacitor and resistor are needed to set the frequency of the internal keyboard multiplexing oscillator. If the key designated 'C' is pressed the generator switches the respective notes of the bottom octave that go to make up the chord of 'C Major' on to the chord element output lines. 'C' would appear at pin 31 being the root note, 'E' at pin 29 as the third and 'G' at pin 28 as the fifth. These three elements

Fig. 3 (below): Pinout details of the AY-5-1317A chord generator.

Fig. 4 (right): Simplified interconnection diagram of the chord generator. This chip operates in conjunction with the AY-5-1315 rhythm generator IC to produce both chords and walking bass effects.

Although we intend to publish reduced scale artwork for the PCB in the next issue, full size etching artwork will be available from the editorial address at the front of the magazine, price 40p. Send a cheque/postal order for 40p together with a large (at least 13" × 11") stamped and self-addressed envelope. Ready-made boards will be available from the Readers PCB service as usual, see ad.





are externally mixed together to produce the chord. If the key designated 'G' were selected the unit would present the notes 'G', 'B' and 'D' at these outputs respectively. If the 'Minor Select' key were depressed at the same time as the chord select, the third element at pin 29 would be changed to a note a semitone lower. If '7th Select' were depressed an extra note would be presented at pin 32. This would be the note a full tone below the root of the chord in question (in both major and minor keys). When this extra note appears it is mixed with the three others producing a very pleasing chord that is frequently used in music notation.

The output designated 'Any Key' is very important. It is made active whenever a key is pressed. It is therefore used to start the rhythm generator in synchronism with a key depression. In the main block diagram this link is anotated 'Rhythm Start'.

Pin 34 is called the 'Special Effects Multiplex Output'. Its effect is not quite so difficult to understand as its name implies. For a given chord selection we can obtain either the root, third, fourth, fifth, sixth or seventh element at this output but naturally only one of these six options at a time. The option is set by the binary code being fed to the control inputs at pins 25, 26 and 27 according to Table 1. If the codes

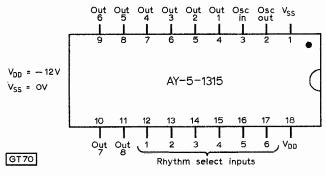


Fig. 5 (above and right): The pinout and simplified connection diagram of the AY-5-1315 rhythm generator. Readers should specify 'pattern 004' when purchasing this IC.

were generated sequentially, it would create an output sequence of the elements comprising the chords. The chord generator chip divides these elements by 1, 2 or 4 so that they fall at least one octave lower than the chord outputs, thus producing a bass note. If the selection were sequenced while a chord was being held it would have the essentials of a walking bass.

Table 1

Pin 27 (B1)	Pin 26 (B2)	Pin 25 (B3)	Note Selected
0	0	0	No change from last note
. 0	0	1	Root
0	1	Ô	Fifth
0	1	1	Third
1	1	1	Seventh
. 1	1	0	Fourth
1	0	1	Sixth

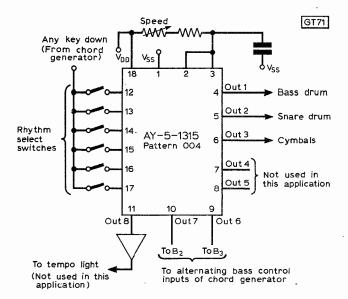
In our application we do not make full use of all the special effect options—we only use two of the control inputs (pins 25 and 26). These are coupled to the rhythm generator which is pre-programmed to change the code in synchronism with the rhythm patterns. This gives an alternating bass which is sounded at the same moment as the bass drum beat.

#### **Vamp**

It should now become easier to understand the details shown in the main block diagram for the accompaniment section. If one assumes that the vamp switch is off, the percussion effect circuits can, for the moment, be ignored. When an accompaniment key is pressed the selected chord elements are presented to the mixer, pass straight through the envelope shaper and on to the tone forming circuits producing two voices 'Flute' and 'String' (Sine and spike respectively). With none of the rhythm select switches closed and the rhythm generator chip outputting the correct code on the bass select lines, there will be a continuous note on the bass output line. This will be the root note of the chord in question.

The bass line from the chord generator is fed to its own envelope shaper and thence to a single tone colour filter. This provides a filtered square wave which sounds reasonably mellow and full. Both the chord and the bass line are mixed before being fed to a volume control and pre-amplifier.

When the Vamp switch is on, the bass note and chord are produced in exactly the same way but do not get through their respective envelope shapers until these receive trigger pulses from the snare drum and bass drum outputs of the rhythm generator. These pulses are short and give a fast attack to the



accompaniment which is followed by a decay set by the sustain circuit within the envelope shapers.

Component values are pre-determined to make the bass note sound like the string of a bass guitar being plucked. Depending on whether flute, string or both are selected, the chords will sound as if played on a piano, banjo or guitar. The bass guitar sounds on the bass drum beat while the chord is struck in synchronism with the snare drum.

When switched to Vamp mode, the accompaniment signals will only appear at the output if the rhythm generator is running; ie if one of the rhythm select switches is on.

#### **Rhythm**

The GI rhythm generator used in this project (Fig 5) is mask programmable so that manufacturers can design their own rhythm patterns. It is not economic to do this for small quantity purchasers so we have settled on one of the standard patterns already being produced by GI. This gives a reasonable selection of six beats, as shown on the figure, but a feature of this circuit is that the rhythms can be mixed by selecting two or more at the same time. Although the chip produces five instrument outputs (plus the alternating bass control signals) we are only using those for bass drum, snare drum and cymbals. These pulses are fed to noise generator circuits which are coupled in a mixer before being fed to their own preamplifier. This, again, has its own volume control.

By having separate volume controls on the melody, accompaniment and drum lines the overall balance may be adjusted to individual preference. It also means that the drums can be reduced to zero level while maintaining operation of the rhythm unit which keeps the vamp going. One can operate the accompaniment and drums section in various modes depending on the set of the instrument tabs. For example one can have straight chords with a static root base note; one can have static chords with an alternating bass and drums or a complete vamp—alternating bass syncopated with the chords and drums.

Next month we shall start on constructional details and describe the various circuits as we go along. If you take delivery of the GI Integrated Circuits before the next part do not remove them from their protective packing unless you take steps to prevent damage from static electricity!

## DESIGN YOUR OWN PROJECTS

#### No.1 LIGHT TRIGGER

#### TOBY BAILEY & BOB WHITAKER

This series describes how to design experimental projects for yourself. Each month we hope to set ourselves a reasonably simple problem, produce a set of specifications to which we want our circuit to conform, and design an appropriate circuit. We will give an account of the possibilities we explored and the thought processes we used to arrive at the final circuit, together with details of how we calculated (or guessed!) component values and why we chose particular component types. Last, but not least, we will provide an honest report of what happened when we built our circuit.

One point requires particular emphasis. These are meant to be one-off projects and the methods used to design the circuits will, in some cases, be more casual in approach than they might be if the circuit were, for instance, being designed for mass-production or in the form of a kit or completed unit. This is not because we want to be condescending or that we think that you might need protecting from some of the complexities. It is simply because this is the method of design used by almost all amateurs of some experience when they are experimenting.

#### Problem

For the first circuit we decided to tackle the well-worn problem of producing a device which will turn on a light when the evening becomes dark enough. This old favourite often appears under names such as "electronic candle" or "automatic nightlight" or some similar title. With a bit of luck we will try to show that designing one of these devices for your own specific requirements is not as daunting a task

as it might appear at first. Of course designing your own circuit from scratch is not as easy as ferreting out a published design in the back issue of a magazine, but that is not the point.

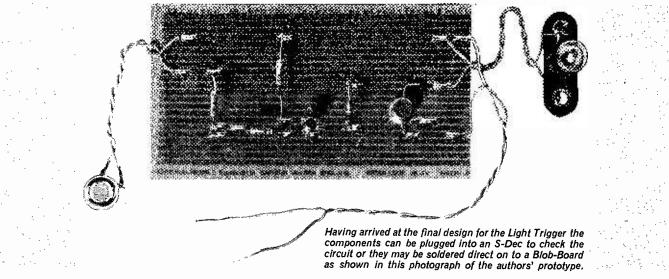
#### **Specification**

So, what exactly are we trying to do? Let's not be overly ambitious, and stick to driving an ordinary torch bulb. The first one we unearthed from the bottom of a drawer has the markings "6V-60mA" still barely decipherable on the side. This seems a pretty reasonable one to use as it doesn't require a terribly unusual power supply anyway. We also need something to sense the light intensity so a light-dependent resistor (LDR) should be just the job. Bob happened to have one of dubious origin attached to a piece of equipment of even more dubious function, from which it was duly severed.

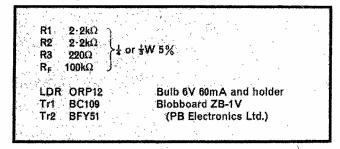
It seemed that the most sensible way of finding out the charactertistics of this LDR was to measure its resistance with a meter. A couple of minutes later we possessed the facts that when it was so dark that we could hardly read the meter the resistance was at least  $100 k\Omega$  and with rather dingy lighting the value decreased to about  $10 k\Omega$ . We decided fairly arbitrarily that the circuit should switch somewhere between these points and that we didn't particularly mind where.

#### Design

It would seem pretty perverse to use anything but 6V for the power supply and, short of using a relay which would be absurdly expensive for our modest



#### **★** components



requirements, we are going to end up by driving the bulb directly from a transistor. So, let's start from that end and see how we get on. The next question that needs to be answered is "does the bulb go in the emitter or the collector?" If we put it in the emitter circuit we shall have to drive the base between about 0.5V and 6V to turn the transistor off and on, whereas if the bulb is in the collector circuit then a much smaller voltage swing is required to get the same effect. As the textbooks more formally put it: "Emitter-follower and common-emitter amplifiers have about the same current gain but the latter has a much higher voltage gain than the former."

So let us accept the extra voltage gain as a bonus, it's free and there does not appear to be any obvious disadvantages associated with our choice of the common-emitter circuit. We now have a circuit like that of Fig.1.

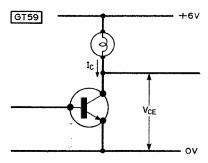


Fig. 1: The use of NPN transistors in the design was determined more by the contents of the junkbox than anything else.

You may have noticed that we have tacitly assumed that we are going to use an NPN transistor. The reasons for this are by no means deep and not entirely unconnected. Firstly, we both have rather more of the NPN type than PNP and, secondly, they tend to be marginally cheaper. Now we need to decide which particular transistor to use, especially since a collector current of 60mA is not small enough to ensure that any transistor we try won't blow up. The highest power dissipation in the transistor is when the collector is at half the supply voltage, i.e. VCE= 3V. This enables us to calculate the dissipation, given by VCE x Ic, and we get  $P=3 \times 60=180$ mW. Now the allowable dissipation for most common small transistors (e.g. BC109, 2N3704 etc.) is around 300mW so we seem to be fairly safe here.

But wait a minute! When a light bulb is just turned on the resistance of the filament is much lower than it is after the filament has had a chance to heat up. This means that the current and hence the dissipation will be rather larger at turn-on. We measured the cold resistance and found that this meant that the peak collector current would be about 150mA, giving a dissipation of 450mW for a

short time. To be on the safe side let us use a BFY51 which can take up to 800mW. Our catalogue tells us that the minimum gain is 40 but neglects to specify the current to which this figure refers. Oh! well, let's say 30 then. This means that we need 2mA at the base of our BFY51 to turn on the light. The stage we have now reached is summarized in Fig.2.

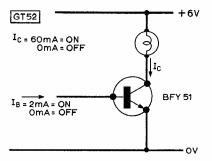


Fig. 2: Already a BFY51 has been chosen in place of a BC109 type because of dissipation problems.

At this point it now becomes abundantly clear that we are going to need another transistor. Since the minimum resistance of the LDR at which the bulb must be turned on is  $10k\Omega$  we would need 20V to pass 2mA through the LDR. So how do we connect the second transistor? It can be either NPN or PNP and can be driven from either the emitter or the collector, giving the four possible circuits shown in Fig.3.

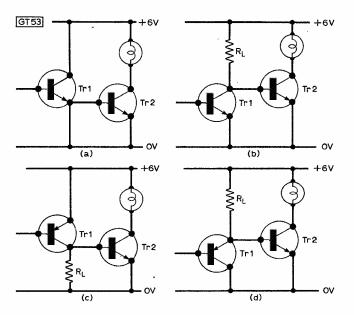


Fig. 3: A second transistor is added but there are four possible ways in which it could be used.

Well, (d) is utterly hopeless since it won't drive the base of Tr2 any lower than around 0.6V so we cannot even guarantee to turn it off. In (c) Tr2 is not going to last very long if Tr1 turns on heavily since Tr1 will then supply a large base current to Tr2. We now have to choose between (a) and (b) both of which use a NPN Tr1. (a) suffers a similar defect to (c) in that we are going to have to be very careful that we do not supply too much base current to Tr1 in order to protect Tr2. Also it is conceivable that when Tr1 is turned off the leakage current might be high enough to upset Tr2, especially if the latter has rather more gain than our conservative estimate.

Circuit (b) seems to be best and we shall now proceed, using this one.

#### **Practical Values**

The next thing to work out is the value of RL. When Tr1 is turned off there will be about 5.5V across RL which then has to supply 2mA to the base of Tr2 (we are ignoring the leakage of Tr1 which should be negligible). For this condition we get  $RL = \frac{5.5}{2} = 2.75 k\Omega$  so let's use a  $2.2k\Omega$  resistor

to be on the safe side. Having fixed this value we can now work out that Trl will have to sink a current of  $\frac{6}{2\cdot 2}=2\cdot 7$ mA in order to turn off Tr2. So we want a nice little low power high gain device for Trl, something like the ubiquitous BC109 should be sufficient, Fig.4.

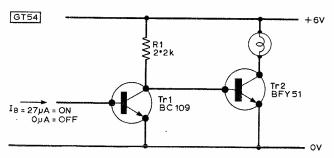


Fig. 4: The value of the load resistor for the BC109 has been decided upon.

The gain of a BC109 is stated as lying in the range 200 to 800. If we assume a value of 100 at worst we're going to need a base current of up to  $27\mu A$  to turn on Tr1. Well, we want the BC109 to turn on when the resistance of the LDR drops to a low enough value so we are going to have to connect the LDR from the base up to the positive power supply line as shown in Fig.5.

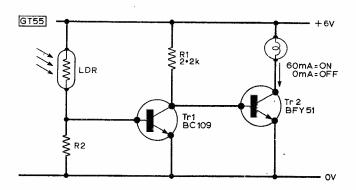


Fig. 5: Safety resistor R2 and the LDR are introduced into the circuit.

Why have we added the extra resistor R2? Well, the LDR with a resistance of say  $100\Omega$  is still conducting  $60\mu A$  (note that in calculating bias resistance it's almost always convenient to ignore the resistance of the base-emitter junction: more correctly the input resistance). Let's assume that if the LDR is at  $100k\Omega$  then we want to pull the base of Tr1 down to a 3V to leave a good safety margin. For this condition R2 has to have a value of at most  $\frac{0.3}{5.7}\times 100 = 5\cdot 3\Omega$  so  $4\cdot 7k\Omega$  seems safe enough. Now if the value of the LDR drops

to  $10k\Omega$  the base voltage will rise to  $\frac{4.7}{4.7}$ 

1.9V. We had better just check that the  $27\mu A$  base current now being drawn by Tr1 does not effect this too drastically. Well, the current being drawn by the bias network of the LDR and R2 is over  $400\mu A$  so the  $27\mu A$  not required by the base is going to make hardly any difference and we have a good margin of error with a base voltage of 1.9V.

#### Moment of truth

Right, now comes the moment of truth when we have to wire up the components and see if we have made any drastic mistakes with the design. After a brief flurry of activity with an S-Dec the circuit was assembled and it worked first time. Well no, to be honest it worked second time, we had a resistor in the wrong hole!

The only problems are that it switches rather slowly if the change in light intensity is not relatively fast and that it has a tendency to get caught in a half on/half off state. What can be done about this state of affairs? The answer is positive feedback and this will have several beneficial effects. It will make the switching faster and it will tend to hold the circuit in the state in which it happens to be. This means that the circuit will not 'jitter' around the switching point. If it becomes dark enough to switch on then it has to get considerably lighter before it will switch off again. This effect is known as 'hysteresis'. How can we achieve this desirable characteristic? Well, the only convenient place from which to take the feedback seems to be at the collector of Tr2, as shown in Fig.6.

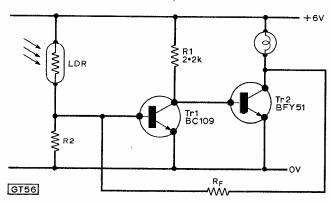
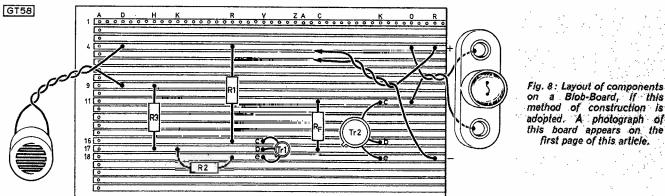


Fig. 6: Positive feedback has been introduced by R<sub>F</sub> which improves the switching characteristics of the circuit.

Is this OK? If the base voltage of Tr1 rises it will turn on further, cutting off Tr2 even more. The collector voltage of Tr2 will hence rise which will tend to cause the base voltage of Tr1 to rise via the feedback resistor. So far, so good then. Oh dear, we shall have to recalculate the value of R2 again as well as that of Rf. As is often the case when choosing the values of two interacting resistors the easiest thing to do is guess the value of one of them and see how it goes. How can we go about guessing a reasonable value for RF then? Looking at the circuit it is apparent that if RF has a much lower value than that of the LDR then the LDR will never be able to change the state of the circuit. If RF is too large its effects will be reduced and we won't get enough hysteresis. Hence we shall choose a value of RF comparable with that of the LDR, say  $100k\Omega$ .



method of construction is adopted. A photograph of this board appears on the first page of this article.

Now when the collector of Tr2 is at 6V RF must not be able to turn on Tr1 without the aid of the LDR. Let us aim to try and get the base voltage down to 0.3V, at least. With an LDR resistence of 100kΩ we have an effective resistance between the base and the 6V line of  $50k\Omega$  ( $100k\Omega + 100k\Omega$  in parallel) so 0.3 we can calculate R2 to be  $\frac{0.3}{6.0-0.3} \times 50 = 2.63 \text{k}\Omega$ , so let's say  $2 \cdot 2k\Omega$ . We have now got to check that

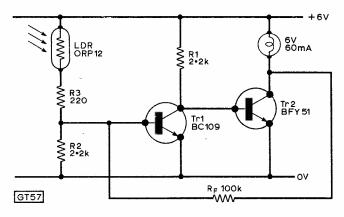


Fig. 7: Final circuit with values of all the components.

when the LDR has a value of 10kΩ the circuit has turned off. If Tr2 is on then RF will effectively be in parallel with R2 and hence the former may be ignored as its resistance is fifty times higher. So, ignoring the base current again, we find that the base voltage of Tr1 is  $\frac{2 \cdot 2}{10 + 2 \cdot 2} \times 6 = 1$ V so this is plenty

to turn on Tr1. We can now draw the final circuit as in Fig.7.

You will see that a  $200\Omega$  resistor has been added in series with the LDR. The truth of the matter is that neither of us noticed until we constructed this stage that there is the possibility that if the LDR resistance drops down to the 100Ω region (in very bright light) then we would have about 6mA of base current in Trl. Whilst little harm is likely to come to our BC109 because of this it seems wasteful and the series resistor will protect Trl if one of us manages to short out the LDR in an aberrant moment.

The circuit now switches cleanly and quickly regardless of the slowness of change of the light intensity and it is impossible to get the output into an intermediate state.

If you wish, you can build the circuit on a PB Electronic's 'blob board' such as the ZBIV as shown in Fig.8 and the photograph.

#### **Postscript**

Well, we have finished now. You may wonder why at each stage we took fairly pessimistic values for the gain of our transistors and chose our resistor values so as to give something in hand over the value we actually calculated. The idea is that all this gives us sufficient in reserve to deal with all the little things we did not consider such as running the transsiistors at voltages and currents different from those at which the gain is specified, resistor tolerances or variation of battery voltage.

We hope that it is going to be one of the main messages of this series. For experimental projects you can ignore all these minor effects providing that you design sensibly with "power in reserve"!

Next month's contribution will describe the design procedure for the circuit of a power supply intended to operate a cassette recorder in a car.

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lov 76	Low Level Battery Indicator	A016	0.40+12
lov 76	Electronic Thermostat	A017	1.30+12
lov 76	Cirtest Probe		
lov 76	Burglar Alarm	A018	0.48+12
	<del>-</del>	A019	0 · 50 + 12
ec 76	Chromachase	A021	5·70+22
an 77	Oscilloscope Calibrator	A023	1 · 25 + 12
an 77	Icelert	A020	1·45÷12
an 77	Polyphon, motor and main boards		7·90+20
	Polyphon, tune disc, blank, SRBP	*800A	0·90+15
	Transistor Checker	A 026	1.18+12
1ar 77	FM Stereo Touch Tuner	D023/4/5	7·50+20
pr 77	Tug 'o' War (set)	A029/030	2.88+12
pr 77	Gas/Smoke Sensor Alarm	A028	0 ⋅ 65 + 12
	2-Way Intercom	D019	1 · 28 + 12
	Protected Battery Charger	A027	2.38+12
	Seekit Metal Locator	A031	3.38+12
	Reverberation Amplifier	A032	2.38+12
uno 77	Versatile AF Generator	A032	_
	Tele-Games		2.38+12
	20W IC Amplifier	D029	3.22+18
	Radio 2 Tuner	A034	1 · 38 + 12
-	Digital Clock Timer	A035	1 ⋅ 68 ⋅ 12
		A036	3.28+12
-	Shoot (Telegames)	D035	1 55 + 15
	Atomic Time Receiver	D036	2·65 + 15
	Morse Code Tutor Cards (SRBP)	A037	4·75+15
	Jubilee Electronic Organ	A038	19·00+75
ept 77	Electronic Car Voltage Regulator	D 037	1 · 25 + 12
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## electronie CAR VOLTAGEREGULATOR

C.GRAYSON B.Sc

In this age of sophisticated electronics in the car the traditional method of regulating the supply voltage by a pair of vibrating contacts with all the associated pitting, arcing and electrical noise, is, to put it mildly, outmoded. Therefore, this simple but reliable electronic regulator was designed.

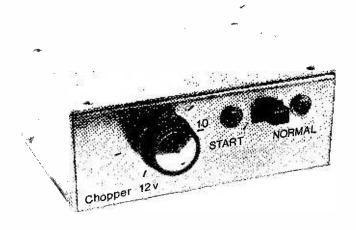
The 'chopper' was originally designed to be a regulator for motorcycles but it can be calibrated for any DC field and armature, dynamo generating system. A CMOS integrated circuit was chosen for the heart of the circuit because of its high immunity to electrical noise, of which there is a great deal in car systems, and its ability to operate at a variety of supply voltages, of which there is precious little on a motorcycle.

#### IC Details

Some readers may be unfamiliar with the integrated circuit used in this project. It is the RCA CD4007A described as a 'dual complementary pair plus inverter' and its internal configuration is shown in Fig. 1. Basically the device consists of three CMOS inverters in various stages of dismemberment. Consider the inverter comprising pl and nl. These are, respectively, a 'p' channel and an 'n' channel enhancement mode MOSFET. These may be considered as a resistive bar of 'p' or 'n' type silicon overlayed with an insulated metal electrode. The potential on the electrode (gate), controls the number of holes or electrons allowed to pass along the bar from source to drain. Thus, each MOSFET is basically a voltage controlled resistor with an  $R_{ds}$  on of about  $500\Omega$ and an  $R_{\rm ds}$  off of greater than ten thousand megohms! Now if pins 13 and 8 are connected together on the CD4007A and a positive voltage is applied to pin 6 then the 'n' channel MOSFET is in the ON state and the 'p' channel is OFF. Thus, the output at pins 13 and 8 is the inverse of the input. In the 'chopper' two inverters are formed with the inputs at pins 6 and 10. The third 'p' channel device is not used but the 'n' channel is used in isolation as a voltage controlled resistor.

#### The Circuit

Now we can consider the circuit Fig. 2, of the 'chopper' in greater detail. Transistor Tr1, diodes D1 and D2, and resistors R1 and R2 form a constant current source of about 8mA which feeds D7 and C1 to give a regulated voltage of about 4V, which is fed to the supply pins of the IC. In the final analysis, this is the reference voltage for the whole system as the



The one control on the Regulator adjusts the voltage while the slide switch enables the system to be started up if the car battery is flat.

voltage-resistance characteristic of the feedback MOSFET is not absolute but relative to the substrate potentials.

The heart of the circuit is a CMOS oscillator with the halves of its cycle separately set in duration. The basic CHOS oscillator is shown in Fig 3. If, say, the output of N2 is high then  $C_{tc}$  becomes charged positive, consequently the input to N2 is high and its output is low, thus  $C_{tc}$  is discharged to ground through  $R_{tc}$ . When the voltage on the input of N1 passes through its transfer point, the output goes high and

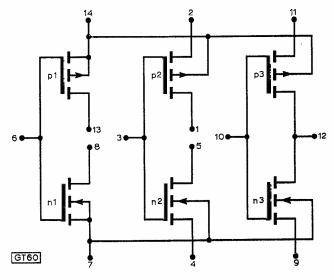
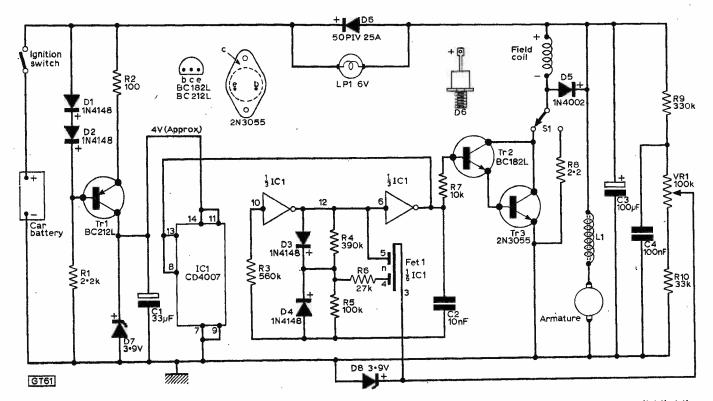


Fig. 1: Basic circuit of the CD4007 IC used in this project as a 'chopper'.



as a result the output of N2 goes low charging  $C_{to}$ . Now  $R_{to}$  provides a charging path for  $C_{to}$  until the transfer point is again passed and the cycle repeats. For an ideal transfer point of 50% of the supply rails the time period of each half of the cycle is about  $0.7R_{to}C_{to}$  but this rises to around  $1.1R_{to}C_{to}$  when  $R_x$  is included to limit the current through the input protection diodes of N1 and to make the frequency of the oscillator relatively free from supply voltage variations.

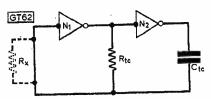


Fig. 3: Basic circuit of the CMOS oscillator, the operation of which is described in the text.

In the voltage regulator the oscillator is used to chop the supply to the field coil of the dynamo. Obviously, the longer the current is allowed to flow in the field coil via Darlington pair Tr2, Tr3, then the higher will be the armature voltage. Accordingly, the OFF period is set at  $1.25 \, \mathrm{ms}$  by C2 charging through R5 when the output of N1 is high and the ON period is variable from  $350 \, \mu \mathrm{s}$  to  $5 \, \mathrm{ms}$  depending on the combination of R4, R6 and the  $D_{ds}$  of FET1.

The action of the circuit is as follows; when the machine is started up there is no voltage fed back by potential dividing chain R9, R10 and VR1. Hence, the  $R_{\rm ds}$  of the FET is almost infinite and the field coil is energised with a 4:1 mark-space ON waveform and the armature voltage rises to feed the vehicle battery, light, etc through power diode D6 (which is necessary to prevent current feeding back from the battery into the armature winding). A proportion of this voltage is fed back to the gate of the 'n' channel FET to lower the total resistance presented to  $C_{\rm to}$  as

Fig. 2: The complete circuit of the Regulator. It is essential that the transistor Tr3 and Diode D6 are insulated from the metal case with the proper insulating kit or washers.

#### \* components

Resistors R1 2 2kt2 R2 18001 R3 550kD		Ω		30k0
pe soon	R6 27kC		R10	33kΩ
P3 SSOLO	R7 10kC		VR1	OOkΩ
R4 390kΩ	R8 2-20	1 5W	linear	oot
. All résistors ≵W i	The second second second		TUDGECOME	200300-0105-2005-2006-00
	(A)			
C1 33µF 16V ele	ctrolytic	C3 100	μF 25 V	electrolytic
C2 10nF paper f	oil	C4 100	int dis	c ceramic
		397.49		
Samiconductors	77.43	01/2055		
TO BOSISE	101	COMO	76	
TE DLIGEL	100 m	SAV DI	A ROW	
DU4 134140	0.7/0	Zacer	9 . D.V. /F	TVRRC9V9)
Semiconductors Tri BC212L Tr2 BC182L D1/4 1N4148 D5 1N4002	0.70		1909.50	
Miscellaneous			2000	
Insulating kit for	2N3066.	Insulation	g wasi	iers for Di
Atuminium case	approx	100 × 70	× 40m	m or large
St. slide switch	DPDT (co	onnect co	ntacts	in paratiel
LP1, lamo 6V 4	0m A. L1.	50 turns	285W	G insulate
Course of County (Control of the Control of the Con	ACCUSED THE SE	facilta t	od. PC	B from PV
copper wire on 1	ummi ais.			

an ON time discharge path. Thus, the field coil is excited for a lower proportion of the time and the voltage tends to fall. The converse is true, and a state of equilibrium is reached and the voltage is stabilised irrespective of engine speed or load. If VR1 is set to tap off less voltage then the regulated voltage at the anode of the power diode will rise to maintain the equilibrium thereby giving a wide range of control. The inductance of the field and armature coils combine to integrate the pulsed drive waveform to give a smooth output voltage.

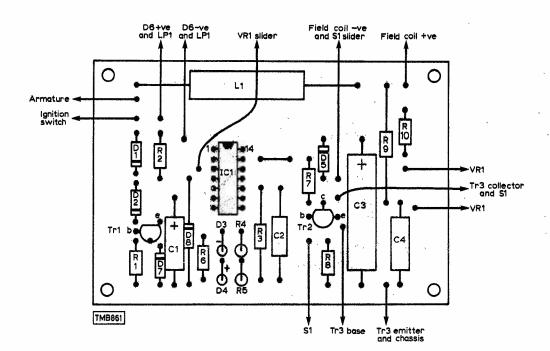


Fig. 5: The components mounted on the PCB are fitted as shown here. Check the lengths of the leads required for fitting the unit to a particular vehicle before soldering them to the board.

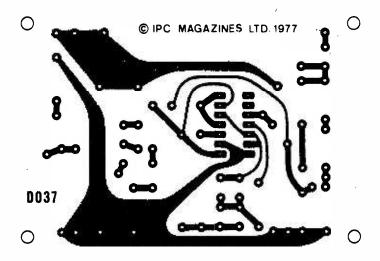


Fig. 4: Foil pattern of the printed circuit board, shown here actual size.

Anyone who has looked at the output of a car dynamo armature will aftest to the apparent lunacy of connecting a delicate CMOS input directly to it. The 'chopper', however, incorporates a second order L-C low-pass filter comprising L1, C3 followed by a first order R-C low pass with R9, C4 to eliminate the hash from the output. Further protection is afforded by D8 which prevents any voltage greater than 4V appearing at the input which would forward bias the internal protection diodes and damage the integrated circuit. Finally, D5 protects the driving transistors from induced high reverse voltages. LP1 is the dash-board discharge lamp which glows when there is a reverse voltage across D6, and S1, R8 provides the facility to start the system up if the battery becomes very flat or disconnected.

#### Construction

Good solid construction is of the greatest importance where the circuit is to be exposed to the hot, dirty and vibrating environment of the road.

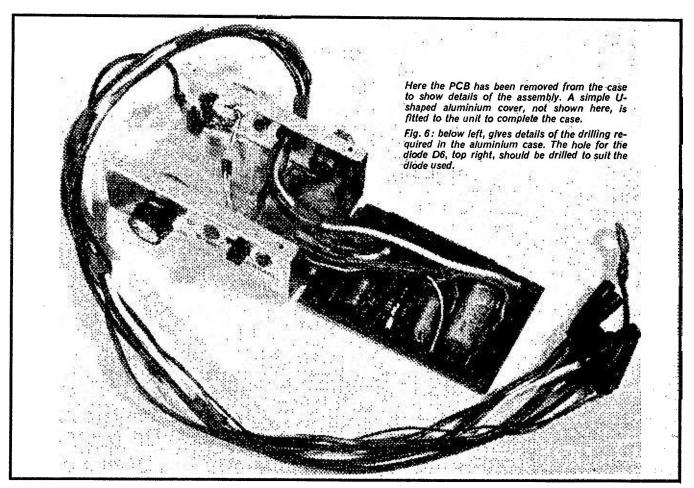
Assembly is greatly eased by the use of the PCB shown in Fig. 4. Coil L1 is formed by winding about 50 turns of 26 SWG insulated wire on a 70mm length of 10mm diameter ferrite rod. The whole assembly may be conveniently secured to the board with epoxy resin. After L1, mount the passive components followed by the semiconductors and finally the integrated circuit. Care should, of course, be taken in handling CMOS. The supply pins should be soldered first with a well-earthed iron followed by the remaining pins. Component layout is shown in Fig. 5.

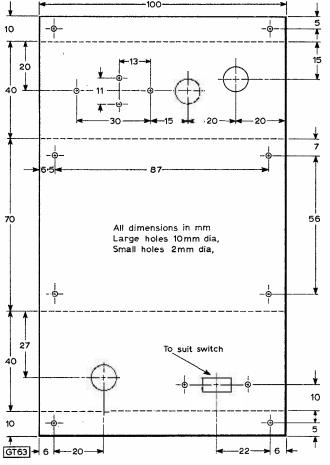
The board is mounted in a standard  $100 \times 70 \times 40$ mm aluminium box. Tr3 and D6 are mounted on the rear face as shown in Fig. 6. S1 and VR1 are similarly mounted on the front face. As the power transistor operates in the switched mode very little power is dissipated and the aluminium serves more as a convenient mounting than a heatsink. The collector of the transistor is electrically isolated from the earthed chassis by a mica washer and bushes. The power diode is similarly mounted through a 10mm hole, again, under all but the most exceptional of circumstances, no heatsink is required. Wiring is quite straightforward but care should be taken to use heavy gauge copper wire in the high current section of the circuit.

#### Setting Up

The usual configuration in the electromechanical regulator situation is for one end of the field coil and one brush of the armature to be connected to the chassis while the other ends (usually marked F and D) go off to the regulator where they are joined by a resistor of some 3 to  $4\Omega$ . Now the earthy end of the field coil must be disconnected and wired into the circuit with the other two terminals as shown in Fig. 2. A little trial and error may be necessary with the values of R9, R10 and VR1 to give the correct range of adjustment and voltage.

As the characteristics of dynamos and CMOS devices vary so much a greater degree of control may be required. To achieve this the value of R4 may be increased to  $1M\Omega$  while R6 is decreased to  $10k\Omega$  to give a 10:1 variation in mark-space. With





the values as shown the range of adjustment is from 10V to 16V on a Lucas type C40-1 generator. For a 6V system reducing R9 to  $100k\Omega$  and R10 to  $10k\Omega$  should prove to be about right.

When adjustments are complete the unit should be securely bolted to some convenient earthed part of the chassis with the leads to the dynamo, ignition switch and charging lamp taken out through a hole in the rear of the case and terminated in standard bullet connectors.

#### **Positive Earth**

Finally, it might be added that a positive earth version of the 'chopper' is quite possible and would merely involve reversing all polarised components except D3 and D4, together with transposing Tr1 and Tr2. A 2N2955 or similar pnp power transistor must be substituted for Tr3 and finally the track to the negative supply pin of the integrated circuit must be broken and the supply voltages transposed.

Now, the voltage feedback to the FET acts on the  $R_{\rm ds}$  in the opposite sense, but the pnp transistors require the complementary logic level to turn them ON. Thus the two phase inversions cancel themselves out and the PCB layout given can still be used. However, with the advent of more electronics in the car many readers will probably wish to do as the author did and take the opportunity of fitting the 'chopper' to convert a positive earth system to the more convenient negative earth. This consists of removing both field and armature coils from the chassis and reconnecting both in the opposite sense, not forgetting to reverse the connections to the vehicle battery and any other polarity sensitive items already fitted.

## -OTLINES

#### A REVIEW OF RECENT DEVELOPMENTS

In general, the author does not have any more information on products than appears in the article.

#### **Measured charge**

One or two items currently launched on the electronics market tend to remind us of the advances made in the field. Solid state relays, for example, have edged their way up the powerhandling capability curve with almost monotonous regularity. I see that one of the very latest to appear (the GB15000 type) can handle a continuous current of 40A with a peak voltage rating of 600V.

Another item is a capacitance meter. Not much bigger than a small hand-held calculator it has a universal socket arrangement for the input(s), a push-to-read button, and a liquid crystal readout. To measure the value of your capacitor, you just poke its leads into the appropriate socket holes and press the button. The unit will give capacitance values from 0·1pF right up to 0·2 Farads. The meter is autoranging in that a LED illuminates to show that the reading is in pF, nF,  $\mu$ F or mF.

#### Viva la weed

It used to be said that the Chinese were "damned clever", but the French seem to be making a bid for this title: would you believe an electronic greenhouse?

Heart of this system is a thing called Firmin which is contained in an area less than  $5 \times 5 \times 8$ in. Firmin carefully measures things like the soil conditions, the ambient light level, evaporation rates etc. It also computes just how thirsty each plant is and then supplies it with just the right amount of water. The system operates from solar cells as a power source.

Firmin has a calculator and a memory plus a section associated with a motor which controls a valve regulating the water flow. It is continually monitoring and comparing the plants' requirements with the environment so it's quite a precise system. It must be nice for the computer in Firmin to receive an order from hollyhock number five, "A large water—and leave the bottle!"

#### Open sesame!

Popular electronics construction journals have often published details of ultrasonic control systems. Those interested in this area will be pleased to hear that a German semiconductor manufacturer is to launch a set of three ICs which will allow hobbyists to make an infra red control system; useful for things like opening and closing garage doors, TV control, light dimmers etc.

The system employs pulse-code modulation and is initiated by simple push buttons. Up to 60 (sixty) command signals are possible from a hand-held, battery-operated transmitter. The receiver can be built into a space about  $3 \times 3 \times 1$  in.

High hopes are held for the infra red system since it has certain advantages over its ultrasonic brother. For example, in the application of opening and closing garage doors the ultrasonic system could not work efficiently from a hand held device inside the car because the "sound" waves would not penetrate the car body etc. The infra red waves, however, can pass through glass and the IR system could be simply pointed through the cars' front windscreen.

Please note: these ICs are not available as yet.

#### **Powerful FET**

The arguments regarding semiconductor devices and valves at UHF still rages, with the semiconductor people edging a little closer each year. A new device just launched is a FET which can generate 1W at 8GHz with a 9.2dB gain. Impressive though this might be for a gallium-arsenide FET, it does have rather a long way to go to compete with the VTS-5753A1. This is a travelling wave tube which has a power gain of 47dB and a peak output at 3.3GHz of 120kW. I have to smile at the valve manufacturer who said of the semiconductor manufacturers of devices at these frequencies, "Ah yes, they make Watt-nots".

#### Get it taped

Video cassette recorders for the home haven't really taken off as yet in the UK, probably on account of the cost. But I am pleased to learn of sensible co-operation between two European manufacturers. Although individually

developed, a close consultation between them during development has resulted in two systems that are compatible. Both will now record over two hours of colour TV programmes without a cassette change. In real terms this long recording time means that an entire football match could be recorded (or even a full length feature film) all on one cassette.

A useful (?) feature on one system is a highly accurate digital clock which allows its owner to select the turn-on and turn-off times of the equipment up to four days ahead of the time that recording is required.

Interesting to note that one manufacturer claims some 90,000 video recorders are in use in West Germany, and over 250,000 in the US.

#### Interested?

Good news for computer enthusiasts—cheaper items are on the way. One of the newer systems costs only \$495 in the US and consists of a 73-key keyboard, 9in CRT, an 8-bit microprocessor, and 4,096 bits of random access memory which operates with Basic programme language. A built-in cassette player allows easy entry of recorded programmes.

Anyone interested in this area of electronics should drop a line to Mike Reeve. He is the secretary of the Amateur Computer Club and you can reach him at 6 Limes Avenue, North Finchley, London N12 8QN. Letters should be written in Basic English!

#### The write beam

Print-out devices are commonly offered on the electronics market, but one just mentioned in Japan seems to be something else. It uses a laser beam to do the "writing" on a photosensitive drum. The letters/characters are made up to a resolution of ten "dots" per millimetre. Printout is at six lines/inch but the twist in the tail is that this printer will printout 10,000 lines every minute! Perhaps it might be called "Print Sprint"?







#### by Eric Dowdeswell G4AR

The Torbay ARS let their hair down at their AGM with no less than eleven complaints on the current operation of the RSGB. Most are long-standing, such as too-technical articles in Radcom and nothing for the newcomer to amateur radio, late and varied posting of Radcom resulting in some readers not being able to take advantage of the 'For Sale' columns. The alleged delay on book orders etc. surprises me in view of the success of the new computer system. However the 1977 President's Committee may come up with some answers in due course, although I doubt it, as the reasons for the complaints are too deepseated to admit of a simple solution. But one of TARS complaints, that the Society's AGM is always held in London instead of being allowed to 'go on tour', should not be difficult to solve!

First item this month must concern Barry Harper of Wednesbury, West Midlands, who was playing around with an old radiogram when he picked up a distress message from an Italian ship in trouble off Rome. Barry alerted his local police who contacted the coastguard authorities who passed a message to Rome. Their coastguards did in fact then locate the fishing vessel in distress. Barry's local press gave him a write-up but, not unusually, got most of the facts wrong! Apparently, Barry is going to rebuild the radiogram and adapt it for SW transmission! And study for a licence to broadcast! Actually, Barry has a No. 19 set that he has been playing with, mostly on 80m, but he's not too happy with the results on SSB.

Brian Harrison in Ore, Hastings continues to find plenty of daytime activity on 10m including KV4 and 8P6, while 20m turned up an HH which is not all that common on any band. In Dringhouses, York, John Hague continues to swot up the code so let's hope we shall be getting some interesting CW logs from you soon, John. Scottish reader John Overton BRS36909 of Glasgow has been lucky enough to enlist the help of two local amateurs GM4CNF and GM4EAW and now sports a BC348 receiver in place of his old CR100. John plans to sit the RAE in December and has not forgotten to keep up with the code. Talking of the code, I hope that the Morse Tutor in the August issue of PW proves of interest, especially in

the clubs, where it would make an interesting project. This month sees the start of a series for those interested in studying for the RAE, which, I hope, includes most of the readers of this column!

John Reynolds, Bridgend, still plays around with his TRF in lieu of anything better. Being still at school limits the cash available. Well, John, be patient, stick to the 'A' level studies which are, by far, the most important item at the moment and perhaps everything will come right 'ere long.

Changes at the West of Scotland ARS where **Tom Wylie** GM4FDM is now the Hon. Sec. Write to him at 38 Rosedale Avenue, Paisley, if you'd like details of their summer programme. Another reminder of the North West Amateur Radio Convention to be held on 17 and 18 September and organised by the University of Lancaster ARS. Contact P. L. Jones c/o Dept of Physics, Univ. of Lancaster, Bailrigg, Lancaster for details.

Peter Page of Epping, Essex is yet another who has returned to the bands after a long break. He has an Eddystone 840 which still worked when the cobwebs were removed! A dipole for 20m suffices for the moment but plans are afoot to improve this situation. John McFadden keeps going in Belfast finding plenty of South Americans on 15m in the evenings. John mentioned W7ALD/MM. I normally refrain from commenting on these maritime mobile calls since by the time it appears in print the ship is thousands of miles away somewhere else or in dock and the operator paid off!

Once a reader of this column gets his/her own licence I don't normally expect to hear from him/her again! Quite understandable, of course, since the first flush of enthusiasm lasts quite a while and every spare moment is spent on the air, not writing letters but QSL's! So it was good to have a note from Jeremy Hinton who now signs G4EZE (was A8962 and G8LIZ) from Newcastle in Staffs. He used the GE special prefix during the Jubilee celebrations and generated a pile-up of Europeans, just for a change. Nice for the G's to get in the act at last! Jeremy uses an FT200 transceiver into a W3DZZ multi-band trap dipole. He's getting the knack of working the DX but sometimes feels he'd like a kW and a multielement beam! Then where would the fun be, OM?? It would all be too easy, like working through repeaters!

Another pair of fellows enjoying the DX by sharing a KW201 receiver between them are Paul Pasquet A9105 and Iain Christie A9201, near Farnham, Surrey. They have another set, built by G4DPP, kept for the BC bands. Back on the amateur bands, they have amassed 160 countries so far this year. Another event on which a reminder would not be out of place

is the RTTY lecture and demonstration by G3MFJ and G3WWF on 16 August by the Wakefield and District Radio Society at Holmfield House. Ring Bob Firth G3WWF on Leeds 825519 for details.

Alan Doherty reports from Portrush, N. Ireland, that he had little response from his appeal for readers in his area to contact him with a view to forming a club. His success with stations in the Pacific has been mostly over the long path in the mornings. He reports poor returns from QSLs sent out but that is not an unusual situation. Personally speaking, I think the average listener to the amateur bands is wasting his time getting QSLs, not to mention the expense. Much better to spend the time studying for the RAE and then to get QSLs that represent much more of an achievement, a two-way QSO!

The Midland ARS is mourning the loss of Bob Palmer G5PP a past President and long established member of the club, giving lectures and organising constructional projects for other members. He was a great contest man and frequently combined holidays with a DX-pedition. This, of course, meant a great interest in mobile rallies which often meant a lot of hard work for him. After his retirement he was able to spend even more time on his hobby. I remember working him myself on many occasions in CW contests and so I know that the Midland ARS has lost a fine operator.

Good news from Paul Barker in Sunderland who says that he is intending to sit the RAE in December next. He is one of many studying at home so I sincerely hope that our news course on the RAE, starting in this issue, will be of some use to him. Paul, you may remember, is a very keen SSTV type and has not entirely deserted the DX bands in search of new countries to log. Just think, Paul, you will soon be able to answer back! Paul was third in his category in the 1st. Worldwide SSTV contest but still feels he did not do as well as he might have done due to QRM. Paul now has an 18AVT vertical multiband aerial which should have gone up by now.

An interesting letter arrived at the last moment from Harry Leeming G3LLL who tells me that, provided there is sufficient support, he will once again be running an RAE course, later in the year. Send your enquiries to the Principal, The College of Technology and Design, Feilden Street, Blackburn.

Late News, page 374.

#### Log extracts

- P. Barker:— 20m AP5HQ C6ANO (Bahamas) KZ5JM TA1MB VP8SM YB6ACV 20m SSTV DK2ZX LZ1MH OK1KO OZ3WP WA2PCY YU1CS 9G1JX
- A. Doherty:— 20m JY25YJ KA6ML KP6AL KP6BD (Kingman Reef) KX6BU TU2EF VP8HZ VR1X YS1ESH 9M8HG 10m A9XBC PY2HY ZP5SD
- P. Pasquet:— 20m PJ3BW TT8SM 3D6BE 6W8LZ 15m VP2GAH VP8PC TR8GB 9V1SW
  - J. Hinton: worked 20m JY25MB KA6SS
- J. McFadden:— 15m CP5AO EA8LD EP2RL HI8XBH HK1ZF KZ5KS LU5MBC PY2FMP PZ1DR VP2LL 6Y5GB
  - C. Page: 20m CT3BO HR3JJR 6Y5KD
- J. Overton:— 20m HC2CG HR1JAG KH6BTD KS6FL
- B. Harrison:— 20m HH2MC OY5NS ZD8RR 15m HM5HB YN7NI 6W8FZ 10m KV4CI YV5ATX 8P6AJ
  - J. Hague:— 15m EP2RR JY4JW 9H4F
  - All reports are SSB unless indicated otherwise.



#### MEDIUM WAVE DX

#### by Charles Molloy

Broadcasts from the Near East present quite a problem to the medium waxe DXer. The time zones of this part of the world are some two to four hours ahead of GMT which means that many of the medium wave stations there have already signed off for the night by the time European QRM starts to clear at 2300. There are a few exceptions. Bagdad is a regular on 760kHz sandwiched between West Germany on 755 and Switzerland on 764 and it can be heard throughout the evening. Another is Kuwait on 1345 which lies between the BBC on 1340 and France on 1349. Others than can be located in this way are Kermanshah, Iran on 895 just below Milan on 899, Ahwaz Iran on 1390 between European channels 1385 and 1394. Sharjah in the United Arab Emirates on 1575, between East Germany on 1570 and Norway on 1578.

An alternative is to listen between 0300 and sunrise when stations can be heard signing on for the day but the best time of all for DXing the Near East is during the month of Ramadan, the ninth month in the Moslem calendar which starts this year on the 16th August. During this period broadcasting goes on for most of the night, many stations staying on the air all night and presenting the listener with the opportunity of hearing DX that otherwise would be difficult or impossible to hear. Listen for Basra on 692; Damman, Saudi Arabia on 885; Antalya, <u>Turkey</u> on 890; Cairo 4 on 940; Damascus 944; Izmir, Turkey on 962; Taiz, Yemen on 1000; Istanbul 1016; Teheran 1088; Radio Oman 1240; Radio Afghanistan 1280; Aleppo, Syria 1313; Kirkuk, Iraq on 1360 and Hyderabad, Pakistan on 1010kHz. Broadcasts from the North African coast should be conspicuous during the late evening. Algeria is on 529, 548, 575 and 980; Tunisia on 584, 629, 719 and 962, Libya on 674, 827, 1124 and 1250 and Morocco on 593, 611, 701, 827 and 1232. On the long waves try for Istanbul on 182, Azilal, Morocco on 209 and Tebessa, Algeria on 251kHz.

"I was wondering if there is any publication available listing the addresses of the lower powered US stations, mainly the 5kW outlets, as I have logged a few" writes Denis Taylor from Preston in Lancashire. No need to worry about station addresses in North America. Use the station call letters and the location as announced over the air. For example Radio WEAN Providence, Rhode Island, USA will certainly find this 5kwatter. Denis uses a Realistic DX160 and MW loop and a few of his latest loggings are La Voz de Antioquia, Colombia on 750; WEAN on 790; All India Radio, Rajkot on 1070; Radio Tiempo, Caracas in Venezuela on 1200; Radio Coro, Venezuela 1210 and WVOJ Jacksonville, Florida on 1320.

A large souvenir QSL card issued by KDKA Pittsburg (1020) mentions that their signal has been heard at some time or other in every country in the world and Denis thinks this must confirm world-wide MW reception as mentioned recently in this column. Worldwide reception depends both on the location

of the listener and on the level of local QRM. In western Europe QRM is the main problem but it is unlikely, QRM permitting, that Hawaii and parts of the Pacific will be logged in the UK in other than freak conditions, as the Great Circle path between the UK and that part of the world passes over the north magnetic pole, a region of high absorption of medium wave signals.

"Freetown on 1205kHz is quite a good signal here around 2300, and I have just sent a report to them" says Roy Patrick whose QTH is in Mackworth near Derby. This station is the medium wave outlet of the Sierra Leone Broadcasting Service and it goes off the air at 2335 daily with a half hour extension on Saturday. Programming is largely in English. This station does QSL and the address is simply Radio Freetown, New England, Freetown, Sierra Leone.

Roy mentions that the long wave Algerian outlet on 251kHz has been off the air for a few weeks but is now back in service and transmits the French programme of RTA. Another item from Roy concerns MEKO 2 which is now testing from a harbour in Libya, presumably with the permission of the Libyan Government. It has been heard in Derby on 773kHz with announcements in English until it goes off at 2300 (0100 local time).

A letter from Paul Griffith, who lives at St James', in Barbados, refers to a MW loop he is constructing. "I ran into a snag in that I am unable to locate a 500pF tuning capacitor in Barbados and have had to order one from the UK" says Paul. This problem has faced some DXers in the UK as well since 500pF variables do not seem to be in demand, other than by constructors of a medium wave loop! Twin-gang capacitors of around 300pF per section are readily available though and one solution to the problem is to use a single section of a twin-gang to tune the HF end of the band and to switch-in the second section in parallel to cover the LF end. There will be some overlap but the additional capacitance when both sections are in use and the vanes closed, will extend the range by only a few kHz, perhaps to 500kHz instead of 540kHz.

The maximum value (vanes closed) of the tuning capacitor is not at all critical. It is the minimum value (vanes open) that is important as it is this amount plus the self-capacitance of the loop and wiring that will determine the highest frequency at which the loop will resonate. For this reason it is not a good idea to use a twin-gang with the two sections in parallel to cover the whole band. If the loop cannot be tuned to a low enough frequency with the vanes closed then the total capacitance is too low and will have to be increased by switching in additional capacitance. This can either be a fixed capacitor or the second section of a twin-gang variable. There is ample scope for the experimenter who can use a twin or triple-gang capacitor from an old radio or a single gang of lower value than 500pF. One design of loop seen by the writer used a 365pF variable, a switch and a 220pF fixed capacitor to cover the band in two sections, thus getting round the difficulty some DXers experience in constructing a loop that will cover the entire range from 540kHz to 1600 kHz.

My original article, 'Medium Wave DXing', which dealt with a practical loop aerial, appeared in PW for April 1970. Back numbers are not available but Eric Dowdeswell, Asst. Editor, is arranging for reprints to be made. Send an SAE to me if you are interested.



#### SHORT WAVE BROADCASTS

by Derek Bell

It is with regret that these paragraphs that I write are the last. Over the many years, since December 1974, that I compiled the Short Wave News I have made many friends and, I trust, have helped and entertained many more. I started writing the column with the firm intention of serving all aspects of our hobby from aerials to QSL card collecting. It was never my intention to be a "Radio Times" section and I hope that the items of news and information that were put in helped to broaden my readers' horizons. Indeed it is to these very readers that I owe a debt of gratitude since, were it not for their letters, there would have been no column to write, so if anyone can employ a redundant typewriter jockey any offers will be gratefully received!

Having got that out of the way we will pass on to this month's letters and start with Roy Patrick from Derby who passes on the information that Radio Canada is now airing a short DX show on Sundays at 2043. Is this a sign that they are at last reinstating their formerly highly active DX section? Roy also tells me that Radio Nacional Brazilia is to cease its international service due to "technical reasons".

While on Latin Americans, Alan Proctor of Bristol has sent an interesting letter on the best verifiers on that continent. He nominates Radio Nacional Bogota and Radio Guaiba, and says that they replied in less than three weeks. Alan does say however that he enclosed two IRCs and, in one case, some Jubilee stamps. These seemed to attract the attention of a staff member who sent a personal reply and Alan is now a firm advocate of the personal touch.

My recent remarks concerning the oldest DXer prompted a reply from Arthur Leir of Belfast. While not, let me hasten to add, a candidate for that title, Arthur has been an addict since 1939 when he started with an Eddystone 504. He has QSL cards dating from pre-war which must be something to see. After a lull Arthur had the urge to re-enter the hobby and has purchased a VEF204, and asks about the various publications on the market. Well, Arthur, Short Wave Magazine is still extant but personally I would buy World Radio and TV Handbook from Argus Books, Station Road, Kings Langley, Herts.

It is one of life's ironies that DX strikes in the strangest places. Andrew Brade, aged 13, has an old domestic Ferguson 391T with a forty-foot long wire hung on the end. Added to this the cord drive to the dial has broken and so Andrew has to rely on verbal IDs but despite this he has two loggings that deserve a mention, Radio Grenada on 15105 at 2015 and this month's star prize for Voice of the Malayan Revolution on 15790 at 0940 and Andrew logged it for a week. It is reported in WRTVH as a 120kW from Hengyang, China, but I have never seen it reported before.

Recently this column arranged for the delivery of a number of spare speakers to a group of bedfast

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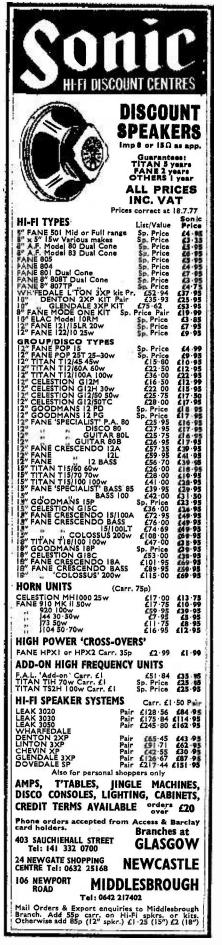
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A C188 A C188K	0·18 0·32	BC183L BC184	0·10* 0·11*	BD410	0.60	BSX19 BSX20	0·16 0·18	2N70 2N11	8 0.15	2N3772 1.90 2N3773 2.10	4014BE 1.00
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AF114 AF115	0 · 20 0 · 20	BC2078 BC212	0 · 11 *	I BDV38	0·60 1·70	BSY55 BSY65	0·33 0·74	2N13	05 0.45	2N4871 0:35* 2N4918 0:60*	4019BE 0.50 4020BE, 1.12
AF116 AF117	0 · 20 0 · 20	BC212L BC213	0.12*	BDY60 BDY61 BDY62	1 - 65	BSY65 BSY76 BSY78	0 30 0 20 0 75	2N13 2N13	07 0.50	2N4919 0.70* 2N4920 0.50*	4021BE 1.03 4022BE 0.95
AF118 AF124	0·50 0·25	BC213L BC214	0.14*	BDY95	1 · 15 2 · 14	BSY78 BSY95A	0.75	2N13 2N13	08 0·60 09 0·60	2N4922 0.58* 2N4923 0.46*	4023BE 0:20
A F125 A F126	0·25 0·25	BC214L	0·15*	BDY96 BF179	4·96	BU105/0	1·80* 2	2N17 2N21	11 0 24	2111020 1 10	4025BE 0 20
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AL102 AL103	1 - 45	BC300 BC301	0·34 0·32	BF181 BF182 BF183	0·30	BU109 BU126	2 · 50* 1 · 60*	2N23 2N24	59 0·14 59A 0·14 83 0·20	2102A-6 3-60 2112A-4 4-75	4028BE 0-91 4029BE 1-10
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BC147	0.09*	BD136	0.39*	BFX86	0.25	ŎČ72	0.22	2N35	70 3.60	25 Pcs 7·50	1 4520BE 1 20



enthusiasts and it is a pleasant duty to thank Mr. and Mrs. Owen of Dollgellan who took the trouble to take these speakers and other components to Bedford.

J. Hilditch of Nantwich has sent the basic brass tacks letter asking "how do I start DXing"? The answer to this is in two parts. The first is decide which sort of DXing takes your fancy, Latin American, North American, Asian, etc. Then, secondly, the thing to do is to decide which of the technical aspects of DXing interests you, aerials, receivers or perhaps it might be propagation. Having decided we can then move on to study these. This can be done by getting hold of a good set of leaflets on your chosen themes. Many stations offer these, generally free of charge, or at worst for the cost of return postage or a few IRCs. The best I know are distributed by Radio Nederland, PO Box 222, Hilversum, Holland, and are free. These leaflets will lead you on to thinking and planning your DX rather than random listening and hoping to catch that elusive station. After all, a fisherman who goes for a particular fish arranges his tackle to suit. It is no good trying for a special station when the darkness path does not suit or your aerial is all wrong!

Some DXers concentrate on collecting QSL cards and in order to do this a knowledge of how to write a letter in a foreign tongue may help and this perhaps will lead to a visit to a reference library. So it can be seen that every aspect of our hobby has a spin-off into some other activity.

With that I fear that I must close down and thank you all for your support and to the readers in all the overseas countries that took the trouble to write, I wish you good DX and best 73s to you and yours.

Reports on SW BC bands should now be sent to Charles Molloy, see box for details.



by Ron Ham BRS15744

Members of the Brighton And District Radio Society were sporting their Jubilee callsigns at the Peacehaven Carnival on June 6th where they gave the visitors an insight to amateur radio. Their 2m station, an FT221 and a 10-element Sky-Beam, installed on the balcony of the Youth Centre was operated periodically by Brian Fenwick GE8BTC, Nigel Hewitt GE8JFT and Alan Baker, GE8LGQ. The Sky-Beam, mounted on a 15ft mast above the balcony, had a mechanical rotator and an electronic beam-heading indicator.

Among the stations worked during the afternoon were F1DZP and a YL operator F1CSW. The group used their own PA system so that visitors could hear the VHF traffic, which took on a special interest when Alan left the station to give a mobile demonstration. A specially prepared leaflet about amateur radio, listing four local clubs, was given to interested people.

During the early hours of the 4th, GE8LGQ took advantage of a lift and worked PE, ON and F on 2m SSB and at 0330 he contacted GE8DJR the station installed at Windsor Castle. Alan worked 36 stations on 2m and the last one was F6DGU/MM operating

a hand-held set on board the Newhaven Ferry 'Vilandry'.

The major sporadic-E disturbances of the 1977 season began on May 25 when strong signals from east-European FM broadcast stations were received between 65 and 73MHz on a dipole feeding my R216 receiver. Similar events, lasting a few hours occurred on May 27 and periodically each day from June 8 to 15th inc., and each time strong vision pulses on Ch.R1 (49.75MHz), warned of the forthcoming event. The frequency range 40 to 75MHz was open all day on the 9th reaching a climax around 1800 when I counted 42 east-European broadcast signals between 65 and 73MHz. However, the most intense disturbance so far occurred around 1800 on the 14th when I heard 46 of these broadcast stations, and for about an hour the sporadic-E extended it's influence to 2m giving G4FUT, G8LIC and G8LZM the opportunity to work YU, YO, HB, LZ and IOSVS in Rome. Sporadic-E upset Band II during the evening of the 7th and 8th, according to Ken Smith BRS20001 Horsham, who heard several continental FM signals between 95 and 100MHz. Ken's report was soon followed by a postcard from "dampest Devon" where John Wilson G8KIS and Richard Lambley G8LAM were on holiday with their families. Richard wrote "We were listening to North Hessary Tor (R4) on my Beolit 600 portable about 1700GMT when the set started being 'captured' by German speaking stations. The whole band was alive with Austrian, Italian and Yugoslav stations until 1800 when it went flat again; all at ground level with a whip aerial and heavy local screening".

Almost every day from May 26th to June 19th there was activity on 10m. Nigel Golds BRS36910 West Chiltington, Sussex heard South-American stations on the 5th in addition to the short skip signals from Scandinavia to Africa which were predominant on most days. Like many of us, Nigel looked first for the International Beacon Project stations hearing 5B4CY on June 1st and 2nd and strong signals from DL0IGI for some period every day. Alan Baker G8LGQ Newhaven, heard the German beacon around midnight on several occasions and on the 8th he just heard the Canadian beacon VE3TEN.

To add to the fun, the sun went "active" on the 4th, Cmdr. Henry Hatfield, Sevenoaks, using his spectrohelioscope, found a sunspot group in the northern hemisphere, a bright plage in the southern hemisphere and a spray of gas on the east limb. Although overcast skies hampered visual observation, Robert Mackenzie, Dover, who projects the sun's image through his telescope, saw the sunspot group on the 8th. John Smith, Cranleigh, reported a marked increase in solar radio noise at 142MHz on the 5th and G8LGQ heard the "seashore" sound on both 2m and 10m indicating the wide frequency coverage of some of the solar bursts.

Both Henry and myself recorded several bursts and a mild solar noise storm at 136MHz on June 5th, 7th, 8th and 9th and the largest individual burst occurred on the 9th and lasted for six minutes. I hope to have reports soon from Kelvin Lake, Edgbaston, Birmingham, who, with the aid of a friend, has just completed a corner reflector interferometer, with a 25 wavelength baseline, operating at 143MHz, for radio astronomy work.

The prevailing high atmospheric pressure,  $30 \cdot 4''$  began to fall on June 2nd and 3rd causing a tropopening toward the north; during this period I received a 569 signal from GB3SUT 432 890MHz, a picture from Lichfield on 189MHz, and several con-

tinental FM stations in Band II with a dipole feeding the receivers in each case. Another tropo-opening began during the evening of the 16th and Barry Ainsworth G8HYN Sompting, Sussex, worked PAO's, F's and GU4EON/M on 2m SSB and at 2330 he contacted G8HVY in Weymouth on 70cms. At the same time, I heard 8 continental FM stations in Band II and a PA0 working EI on 2m. GB3SUT came up at 2130 and was frequently heard in Sussex until the small hours of the 20th. During the afternoon of the 19th G8LGQ had a contact with G2BAR in Bristol and at 1816 he worked GW4FPX/P in Gwent. As if these were not enough to satisfy a VHF man's ego for one day, Alan heard a very weak GM pop up from the noise around 2000 and he stuck on that frequency until 2315 when, for the first time from his home in Newhaven, he worked GM6UW/P just north of Aberdeen, a distance of 560 miles which is not bad going on 2m!

Owing to the prevailing sporadic-E disturbance to the 4m band I was unable to make a satisfactory observation of the June Lyrid meteor shower which reached it's peak on the 15th. However, the next event to look out for are the Perseids which reach their maximum on August 12th. I would also like to remind you all about the 4m Open Contest on August 13th/14th and the 2m Open Contest on September 3rd/4th and in both events SWLs are welcome to

take part.

Ivan Davies G3IZD took advantage of the good conditions and worked OZ1OQ and ON6AG on 2m SSB, from Arundel, Sussex during the evening mobile rally organised by Worthing and District ARC on June 21st. Among the visitors to the rally was Ernie Hoare G8BDJ from Brighton who learned recently that he

Reports on the various bands are welcome and should be sent direct, by the 15th of the month, to:-

AMATEUR BANDS Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey KT21 2TW. Logs by bands, each in alphabetical order.

MEDIUM and SW BANDS Charles Molloy G8BUS, 132 Segars Lane, Southport, PR8 3JG. Reports for both bands must be kept separate.

VHF BANDS Ron Ham BRS15744, Faraday, Grey-friars, Storrington, Sussex RH20 4HE.

was the winner of the 3cm section of the IARU Region 1 contest held last October. Ernie told me that he hopes to get going on 24GHz soon.

Thanks again to all of you who send me reports, I really do look forward to hearing from you

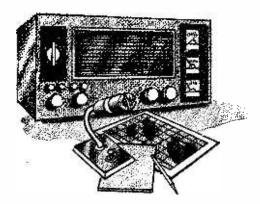
and putting this feature together.

#### Amateur Bands-continued

You won't be able to make the next meeting of the Wessex AR Group, on 5 August, because it's too late! But because it will be a very interesting talk perhaps G6MB and G4AMW might consider repeating it, if you ask them nicely! Title is 'Problems encountered in building and operating HF transceivers using the SL600 IC's. You will be able to make the following date on 19 August when Power Supply Units will be discussed followed by a talk on studying for the RAE. Details from Geoff Cole G4EMN, 6 St. Anthonys Road, Bournemouth (Tel: 20027).

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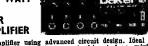
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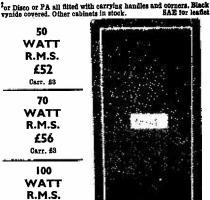
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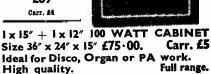
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Š	DL92	·45	ECC80		EY81 .45 EY83 .60	MX40 1 00	PY81 50	U408 -90 U404 -75	ACY28	21	OA79	ii
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10	DM70	-80	ECF82	·šŏ	EY88 .55	P61 -60	PY83 50 PY88 60	U4020 -75	AD161	-58	OA85	11
0	DM71	1.75	ECF86	-80	RY91 .50	PABC80 45	PY500A1-85	VP28 -65	AD162	-98	OA90	-14
0	DW4	1·15 2·00	ECH85		EY500 1.45 EZ35 .45	PC86 -68	PY800 50	VP41 -90 VR105 -50	AF114 AF115	-80	OA91 OA95	11
iŏ	DY51 DY87/6	45	ECH42 ECH81	40	EZ40 -52	PC88 -62	PY801 -50	VR150 -75	AF117	.23	OC44	19
5	DY802	50	ECH88		EZ41 -52	PC92 -55 PC95 1-00	PZ80 50	VU111 1.00	AF121	-85	OC48	-18
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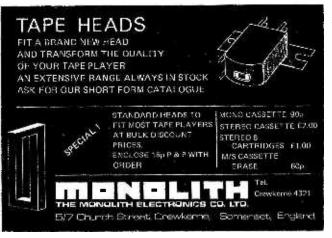
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AA119 0:20 AAY30 0:13	BD181 1:38 BD182 1:48	OC24 3.50 OC25 0.90 OC26 0.90	2N2222 0-25 2N2223 2-75	7437 7438
AAY32 0.15	BD237 0-80 BD238 0-85	OC28 2.00	2N2358 0-17 2N2359A 0-21	7440 7441AN
AAZ15 0:31 AAZ17 0:25	BDX10 6:75 BDX32 2:25	OC29 2:00 OC35 2:50	2N2484 0:21 2N2646 0:56	7442 7447AN
AC107 0:75 AC125 0:30	1 PLOVED 0.7%	OC36 1-58 OC41 0-50 OC42 0-50	2N2904 0:35 2N2905 0:35 2N2906 0:25	7450 7451
AC126 0:25 AC127 0:25	BF115 0:39 BF162 0:25	OC42 0:50 OC43 1:50	2N2907 0.21	7453 7454
AC128 0:25 AC141 6:26	BF152 0:25 BF153 0:25 BF154 0:25	OC43 1-50 OC44 0-50 OC45 0-50	*2N2924 6-15	7460 7470
AC141K 0-30 AC142 0-20	BF160 0:35	OC71 0:45 OC72 0:45	*2N2926 0-13 2N3053 0-25	747 <u>2</u> 7473
AC142 0.26 AC142K 0.25 AC176 0.29		OC71 0.45 OC72 0.45 OC73 1.00 OC74 0.50	2N3054 0:50 2N3055 0:65	7474 7475
AC187 0.25 AC188 0.25	BF173 0-39 BF177 0-38 BF178 0-45	OC76 9-50	2N3440 8:80 2N3441 8:88	7476 7480
AC187 0-25 AC188 0-25 AC177 0-68 AC178 0-65 AC178 0-65 AC178 0-65 AC178 0-65 AC178 0-65	BF179 0-48 BF180 0-45 BF181 0-45	OC77 1:20 OC81 9:75	2N3525 6-90	7482 7483
ACY18 0:65 ACY20 0:65	2 E400 A.46	OC81Z 1 00 OC82 0 75 OC83 0 55	2N3614 1-20 *2N3702 0-15	7484 7486
ACY21 0.65 ACY39 £1.00 AD149 0.70	BF183 0-45 BF184 6-39 BF185 0-37	OC83 0-55 OC84 0-60	*2N3703 8:15	7490 7491 AN
AD161 0.75	*BF194 0-12	OC122 1:50 OC123 1:55	*2N3706 0·15 *2N3706 0·16	7492 7493
AD162 0.78 AF106 0.45	*BF196 0-13	OC140 1.95 OC141 2.25	*2N3707 0:18 *2N3708 0:14	7494 7495
AF114 6:25 AF115 6:25 AF116 0:25	*BF197 0·14 BF200 0·32 *BF224 0·20	OC141 2:25 OC170 9:60	*2N3709 0-15	7496 7497
AF117 0:25	*BF244 0·35	OC170 0-86 OC171 0-88 OC200 1-60 OC201 1-50	2N3771 1:60	74100 74107
AF139 8:40 AF188 1:50	BF257 0:37 BF258 0:42 BF259 0:45	OC202 1.25	2N3773 2·65	74109 74110
AF239 0 45 AFZ11 2 75 AFZ12 2 75	*RF336 0.56	OC203 1.25 OC204 1.25 OC205 1.75 OC206 1.75	*2N3819 0:36 *2N3820 0:46	74111 74116
ASY26 0-45	*BF338 0-55	OC205 1.75 OC206 1.75 OC207 1.25	*2N3823 0.60 2N3866 1.00 *2N3904 0.21	74118 74119
	BECOS 1.38	OCP71 1-25	*2N3905 D-22	74120 74121
ASZ16 1.25 ASZ17 1.25 ASZ20 0.75	*BFS98 0-25 BFW10 0-96	*R2008B 2:25 *R2009 2:25	*2N3906 0-22 *2N4058 8-20	74122 74123
ASZ21 1.50	BFW11 0:90 BFX84 0:38	*R2010B 2:25 T1C44 0:36	*2N4059 0·15 *2N4060 0·29	74125 74126
*AU113 1.70 *AUY10 1.70 *BA145 0.15	BFX85 0:41	T1C226D 1-39 T1L209 0-25	*2N4061 0·17 *2N4062 0·18 *2N4124 0·17	74128 74132
*BA148 0·15 BA154 0·10	BFX88 0-32	*T1P29A 0.59 *T1P30A 0.60	*2N4126 0.17	74136 74141
BA155 0·12 BA156 0·13	BFY50 0.28 BFY51 0.26 BFY52 0.26 BFY64 0.30	T1P31A 0-62 T1P32A 0-75	*2N4286 0·20 *2N4288 0·25 52N4289 0·25	74142 74143
	BFY90 1.32	T1P33A 1.00 T1P34A 1.20	*2N5457 0.35	74144 74145 74147
BAX16 0.07	BSX19 0-34 BSX20 0-34 BSX21 0-32	T1P41A 0:70	*2N5458 0-35 *2N5459 0-35	74148
BC107 0-12 BC108 0-12 BC109 0-13	BSX21 6-32 BT106 1-25	T102055 1-00	3N125 1.75 3N141 0.85	74150 74151 74154
*BC113 0:15 *BC114 0:18	BTY79/400R 3·19	*T1S43 0.35	7400 0 · 20 7401 0 · 20	74155 74156
*BC115 0 19 *BC116 0 19 *BC117 0 22	*BU205 2-25 *BU206 2-25	*Z\$140 0.25 *Z\$170 0.12 *Z\$178 0.54	7402 0 20 7403 0 20	74157
*BC118 0-16	*BU208 2-50 BY100 0-45		7404 0·26 7405 0·23	74159 74170 74172
*BC125 0-18 *BC126 0-25 *BC135 0-15	BY100 0.45 BY126 0.14 BY127 0.15	*ZS271 0-56 *ZTX107 0-11 *ZTX108 0-12 *ZTX300 0-12 *ZTX300 0-12 *ZTX300 0-12 *ZTX300 0-17 *ZTX302 0-17 *ZTX303 0-17 *ZTX304 0-19 *ZTX311 0-19	7406 0.55	74173 74174
*BC136 0.19	BZX61 0.20 Series	*ZTX109 0:12 *ZTX300 0:12	7407 0·55 7408 0·28 7409 0·28	74175 74176
BC137 0-16 BC147 0-10	BZY.88 0·13 Series	*ZTX300 0:12 *ZTX301 0:13 *ZTX302 0:17	7410 8·20 7412 0·26	74178 74179
*BC148 0-10 *BC149 0-13	CRS1/05 0-45 CRS1/40 0-60	*ZTX303 0 17 *ZTX304 0 19	7413 0·45 7416 0·40	74180
*BC149 0-13 *BC157 0-12 *BC158 0-11	CRS1/40 0 60 CRS3/05 0 45 CRS3/40 0 75	*7TX314 0.20	7417 6-40 7420 0-29	74190 74191 74192
*BC159 0:13 *BC167 0:13 *BC170 0:16	CRS3/60 0 90 GEX66 1 50 GEX541 1 75	*ZTX500 0:13 *ZTX501 0:14	7422 0·25 7423 0·35	74193 74194
"BC171: 0:14	GJ3M 0.75	*ZTX502 0·16 *ZTX503 0·17	7425 0-35 7427 6-35 7428 0-50	74195 74196 74197
*BC172 0:13 *BC173 0:15 BC177 0:19	GJ5M 0.75 GJ7M 0.75 GM0378A1.59	*ZTX503 0.17 *ZTX504 0.20 *ZTX531 0.20 *ZTX550 0.16	7430 0.20	74197 74198 74199
BC178 0-18	*KS100A 0-40 MJE340 0-58	1N914 9:07	7432 0·36 7433 0·37	74199 *76013N
*BC182 0-11 *BC183 0-11 *BC184 0-12	MJF370 A-85	1N916 0.07 1N4001 0.06 1N4002 0.07		
*BC184 0:12 *BC212 0:14	MJE371 0-81 MJE520 0-65 MJE521 0-75	1N4003 0:08 1N4004 0:09	DIL	8 pin 14 pin
*BC213 0-14 *BC214 0-17	MJE2955 1 25 MJE3055 0 75	1N4005 0-13	SOCKETS	14 pin 16 pin
*BC237 0.17 *BC238 0.12	*MPF102 0-30 *MPF103 0-30	1N4007 0·15 1N4009 0·15	-	
BC301 0·45 BC303 0·60	*MPF104 0:30	1N4148 0-07 1N5400 0-14 1N5401 0-16		
BC303 0.60 *BC307 0.20 *BC308 0.18 *BC327 0.22 *BC328 0.18	*MPSA06	1544 0-06		
*BC327 0·22 *BC328 0·18	*MPSA56	15920 0:08 15921 0:08		
*BC337 0-19 *BC338 0-18	*MPSU01 0-32 *MPSU06 0-40 *MPSU56 0-45	2G301 1.00 2G302 1.00		
BCY30 1.00 BCY31 1.00	NKT401 2:00	SG306 1.10 2N404 0.60 2N696 0.25		
BCY32 1.00 BCY33 0.90 BCY34 0.90	NKT403 1.73 NKT404 1.73 NE555 0.45 OA5 0.75	2N696 0.25 2N697 0.16 2N698 0.30		
BCY31 1.00 BCY32 1.00 BCY33 0.90 BCY34 0.90 BCY39 3.00 BCY40 1.25 BCY42 0.30	OA5 0.75 OA7 0.55	2N705 0-80 2N706 0-12		
BCY42 0.30 BCY43 0.32	OA10 0.55 OA47 0.14	2N708 0 21 2N930 0 26		
BCY43 0-32 BCY58 0-23 BCY70 0-18	OA70 0:30 OA79 0:38	2N1131 0 26 2N1132 6 26		
BCY71 0-22	OA81 0-30	2N1302 0-37		₹;

8 pin 15p 4 pin **15p** 16 pin **17**p

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2N1131 2N1132 2N1302 2N1303 2N1304 2N1305 2N1307 2N1308 2N1307 2N1308 2N1671 2N1693 2N2147 2N2147 2N2218 2N2218 2N2219 2N2220 2N2221

OA10 OA47 OA70 OA79 OA85 OA90 OA91 OA200 OA200 OA211 OAZ200 OAZ200 OAZ200 OAZ200 OAZ200 OAZ200 OAZ200 OAZ200 OAZ200

0-22 0-17 1-50 0-80 1-50 1-50 1-50 0-35 0-36 0-37 0-43 0-47 2-00

BCY71 BCY72 BCZ11

BD115 BD121 BD123 BD124 BD131 BD132 \*BD135 \*BD136 \*BD137 \*BD138 \*BD138 \*BD139 \*BD140 BD144

0.554 0.39 0.39 0.39 0.39 0.11 0.75 0.65 0.65 0.65 0.65 0.65 0.22 0.22 0.25

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22pf to 47000pf 3p. Mylar Capacitors 100V 001,
002. 005 4p. 01. 02. 025 4½p. Polyester Capacitors 250V 66 01 to 'Imf 5½p. 15, 22. '33mf 7p.
47mf 1½p. Electrolytics 50V 47, 1, 2mf 5p. 23 V 5,
10mf 5p. 16V 22. 33, 47mf 6p. 100mf 7p. 220, 330m
9p. 470mf 1½p. 100mf 18p. Zaner Diodes 400mW
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7407 49p	LINEAR I.Cs AY-1-212 CA3028A CA3046 CA3046 CA3048 CA3085 CA3085 CA3085C CA3090A CA3090	480 p 112 p 250 p 270 p 270 p 270 p 265 p 465 p 465 p 112 p 115 p 115 p 116 p 120 p	NE556 NE5618 NE5628 NE5658 NE5656 NE5656 NE5656 NE5657 SN72710N SN75003N SN75003N SN75003N SN75003N SN75003N SN75003N SN75023N SN7502N SN7502N SN7502N SN7502N SN7502N SN7502N SN7502N SN7502N SN7502N SN	100p 450p 450p 200p 54p 275p 275p 175p 175p 175p 175p 275p 275p 275p 275p 275p 275p 275p 2	AD149 45p AD149 45p AD149 42p AD149 22p AD149 22p AD149 22p AD149 22p AD149 40p AD147	T P31C 48p T P32C 85p T P32C 85p T P33C 124p T P33C 124p T P34C 124p T P34C 124p T P36C 242p T P36C 242p T P36C 36p T P41C 84p T P41
7444 120p 74172 750p 7445 108p 74173 180p 7446 108p 74173 180p 74476 108p 74174 130p 74476 18p 74176 130p 7450 18p 74176 130p 7451 18p 74180 120p 7453 18p 74181 324p 7454 18p 74182 130p 7460 18p 74182 190p 7460 18p 74185 190p 74760 38p 74185 190p	VOLTAGE REC Plastic—T0220— 1 Amp + ve 5V 7808 12V 7812 15V 7815 18V 7815 24V 7824 100mA — T092	3ULATORS -3 Terminals 130p 130p 130p 130p 150p	5-Fixed 7	215p 215p 215p 215p 215p 215p	BC212 14p BC213 12p BC214 17p BC461 40p BC478 32p BC4770 20p BCY71 24p BD124 140p BD124 63p	201307 75p 201613 22p 201711 22p 201893 32p 202102 60p 202160 120p 202219 22p 202229 22p 202389 15p
7472 32P 74190 188p 7473 38p 74191 180p 7474 37p 74192 130p 7475 48p 74193 130p 7476 37p 74194 180p 7480 54p 74195 110p 7481 108p 74196 130p 7482 80p 74197 130p	100MA	70p 70p 70p 70p 1A 5V 8A 5V 1 100mA +1 1 100mA ±1 m ±8V to	78L05 	80p 80p 80p 275p 300p	BD132 67p BD135 554p BD136 55p BD149 66p BD149 60p BD149 625p BF115 24p BF167 25p BF170 25p BF173 27p	2N2484 32p 2N2646 52p 2N2604/A 22p 2N2606/A 22p 2N2606/A 22p 2N2606/A 22p 2N2606/A 25p 2N2606/A 25p 2N2606/A 25p 2N2606/A 25p
7482 \$80p 74167 1880 7483 \$9p 74167 1880 7484 120p 74168 270p 7485 120p 7489 370p 7489 340p 7489 340p 7489 340p 7489 340p 7489 340p 74878 320p 7489 38p 74878 320p 7489 38p 74878 320p 7489 38p 74828 270p 7489 38p 74828 270p 7489 39p 74828 270p 7489 39p 74828 270p 7489 38p 74828 270p 7489 39p 74828 270p 7489 30p 74828 478p 7484 30p 74828 478p	723 14 pin DiL LM317T TO220 TL430 TO92 100 78MGT2C 4 pin i	2V-37V 150n 1A-2V-37V IMA 2V-30V DIL 5V-30V I	nA 500mA	120p 150p 45p 328p 75p 148p	BF177 25p BF178 30p BF180 35p BF180 35p BF181 35p BF184 24p BF185 24p BF185 11p	2N3055 65p 2N3439 75p 2N3442 151p 2N3655 34p 2N3702 14p 2N3704 14p 2N3705 14p 2N3706 14p 2N3706 14p 2N3708 14p
7485 75p 74290 660p 7487 800 74293 660p 7497 200p 74298 220p 74100 140p 74395 160p 74104 75p 74395 160p 74104 75p 74397 160p 74107 36p 74397 220p 74107 60p 74390 220p 74109 60p 74390 250p	OPTO-ELECTF OCP 71 130p ORP 12 75p ORP 60 90p ORP 61 90p 2N5777 48p DISPLAYS 3015F Minitror DL704 Red/Gre DL704 Red/Gre	n 175p een 160p	TIL209 Red TIL211 Green TIL32 Infrared 0.2" Red Green Yellow FND507 Red TIL321 Red	14p 36p 81p 16p 20p 36p 130p 130p	BF197 19p BF204 40p BF244B 34p BF256B 60p BF256 34p BF258 38p BF259 48p BF337 32p BF739 34p BFR40 34p	2N3709 14p 2N3707 14p 2N3711 14p 2N3773 320p 2N3819 50p 2N3820 50p 2N3820 70p 2N3823 70p 2N3866 97p 2N3803 18p 2N3904 22p
CMOS ICs 4000 21p 4040 150p 4001 21p 4042 97p 4002 21p 4043 100p	DL747 Red/Gre DRIVERS: 75	een 250p 491 84p 492 104p	TiL322 Red 9368PC	130p 216p	BFR41 34p BFR79 34p BFR80 34p BFR81 34p BFR88 37p BFR88 37p	2N390b 22p 2N3906 22p 2N4058 19p 2N4060 19p 2N4123/4 22p ZN4125/6 22p
4006   127p   4046   150p   4007   21p   4047   120p   4009   67p   4049   64p   4011   21p   4050   58p   4013   55p   4055   140p   4014   90p   4056   145p   4015   90p   4060   130p   4016   54p   4016   54p	SCR-THYRIST 1A 50V TO5 1A 400V TO5 3A 400V Stud 7A 400V TO22 16A 400V TO22 16A 600V TO22	0 0 20	65p 2N3525 95p 2N4444 120p 2N4444 180p DIODE: 270p BY127	12p	BFX30 34p BFX84 30p BFX85 30p BFX86 30p BFX87 30p BFX88 30p BFX88 20p BFY50 22p BFY50 22p	2N4289 24p 2N4401 34p 2N4403 34p 2N4403 97p 2N4871 60p 2N5089 34p 2N5179 75p 2N5296 58p 2N5296 62p
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7402	16	7481	114	74165	120	4028	95	4098	110	CA3023	170p	NE555*	36p
7403	18	7482	82 95	74166 74167	161	4029 4030	109	4099 4502	120	CA3028* CA3028A*	95p 98p	NE556DB* NE560*	90p 325p
7404	20	7483 7484	85	74172	570	4031	230	4507	55	CA3035	140p	NE561*	410p
7405	44	7485	125	74173	175	4032	100	4510	135	CA3036	180p	NE562B*	410p
7407	44	7486	38	74174	173	4033	145	4511	175	CA3043	190p		185p
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7417	39	7497	262	74188 74190	856	4042	85	4539 4553	110 449	CA3123	200 p	SN76003N	240p
7420	18	74100	135	74191	184	4043	100	4585	106	CA3130*	94p	SN76013N	240P
7421	28	74104	48	74192	128	4044	95	4000	100	CA3140 ICL8038CC	95p	SN76023N SN76033N	140p 230p
7422 7423	24 36	74105	48	74193	120	4045 4046	145 135	LINEAR	ICIE	LM300H	170p	SN76115N	215p
7425	25	74107 74109	84	74194	120	4047	99	702	75p	LM301AP	39p	SN76227N	175p
7426	38	74110	54	74195	85	4048	58	709C 8 pi	n 35 p	LM301AT	49p	TAA550	50p
7427	38	74111	78	74196	118	4049	60	709C 14 p	in 14p	LM304	240p	TAA621A	
7428	48	74116	108	74197	118	4050	55	709C TO5		LM305H	4 40-		228p
7430	18	74118	115	74198	248	4051	95	710* 723* 14 pl	48p	LM308T LM309H	140p 140p	TAA661A	155p
7432	28	74120	195			4052 4053	95 95	723* TO5	n 45p	LM319H	205p	TAA700 TAA960	353p 300p
7433 7437	30	74121 74122	34	CMOS	5 *	4054	120	741 8 pin	22p	LM318S	195p	TAD100	150p
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7441	74	74126	85	4001	17	4059	480	747C 14 p 748C 8 pli	in 70p	LM381	170p	TBA540	215p
7442	.74	74128	81	4002	.17	4060	115	748C 8 pli		LM382	125p	TBA540Q	220p
7443	130	74132	75	4006	115	4063 4066	110 65	748 TO5 753 8 pin	45p 150p	LM3900 LM3909N	65p 70p	TBA550Q TBA641-A	355p
7444 7445	122	74141	72 315	4007	18 97	4067	380	78L82AW	C* 51p	M252AA*	850p	IDAUTITA	250p
7446	118	74142 74143	314	4009	60	4068	22	AY-1-021		M253*	950p	TBA641 B	
7447	82	74144	314	4010	60	4069	22	AY-1-5050		MC724*	175p		250p
7448	88	74145	90	4011	19	4070	60	AY-1-505		MC663		TBA641 B1	
7450	17	74147	275	4012	18	4071	23	AY-1-1313		MC845			250p
7451	17	74148	180	4013	55 104	4072 4073	23 23	AY-1-6721 AY-3-8500		MC1303 MC1304P	145p 260p	TBA651 TBA800	180p
7453	17	74150	128	4014 4015	99	4075	24	AY-5-855		MC1310P	185p	TBA810S	100p
7454	17	74151	79	4016	55	4076	183	AY-5-122		MC1312PQ	195 p	TBA820	80p
	17	74153	82	4017	104	4077	60		349p	MC1458P*	90p	TBA920Q	350p
7460		74154	150	4018	105	4078	22	AY-5-1230	0* 490p	MC1496L	101 p	TCA270Q	220p
7470	32	74155	76	4019	60	4081	22	AY-5-131		MC1710CG	79p	TCA270SC	220p
7472	28	74156	80	4020	125	4082	23	AY-5-131		MEM780		TDA1412	
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7474	36	74189 74180	225 118	4022	20					MK50253*	550p	ZN414 ZN424E	130p
7475	-	1 /4100	110	4020	e.u	1 4009	100	A 1 -0-400	, esop	· INITIONEGO	Joop		, och

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5A400V 7A400V 8A400V BT106 C106D

TIC44 TIC45

120p 125p 150p 150p

191p

10x41x3" 12x5x3" 12x8x3"

	OPTO ELE	CTRO	NICS*		SWITCHES	*
S	LEDS + CI		7 Segment Display	ys	TOGGLE 2A	
	TIL209 Red		5-LT-01	460p		
	TIL211 Grn		TIL312-3" C.An	125p		26 p
le)	TIL220 Red		TIL313 -3" C.Cth	125p		35p
	"2" Red		TIL321 -5" C.An	140p	4 pole on off	58p
	Yellow, Gr	een,	TIL322 .5" C.Cth	140p		
RS	Amber OCP70	23 P	DL704 -3" C.Cth DL707 -3" C.Anod	99p	SUB-MIN	
	ORP12	40p	DL747 6" An	99p 180p	TOGGLE	
65p	2N5777		MAN3610 C. An			
58p	OPTO		MAN3640 C. Cath	260p	SP changeove	
65p	ISOLATO		MINITRON 3015F	240p	01 01 011 011	
65p	TIL111/2	105p		2.40	<b>DPDT 6 tags</b>	780
-	TIL114		DRIVERS		DPDT Centre	off
190p	TIL117		75491/75492	90p		92p
	MOLTACE	DECII	ATORCE		<b>DPDT Biased</b>	115p
205p	VOLTAGE		Plastic Case			
	723 TO5	43P	78L82AWC -1A 8V	/ 51p	SLIDE 250V	
	TBA625B		MC7805 1A 5V		1A DP	14p
113p	TO3 Can T		MC7812 1A 12V			
140p	1A 5V		MC7815 1A 15V	135p	1A DP C O	16p
166p	1A 12V		FT8N24 1A 24V	140p	27 0	13p
195p 1 200p	1A 15V		Variable Type	. 700	4 pole 2-way	24 p
245p	1A 18V		LM325N ± 15V	240p		
398p	LM309K		LM326N ± 12V	240p	PUSH BUT	TON
99p	LM323K		LM327N ± 5 -12	260p		
104p	MVR5		LM317K 20W 1-5A		Shiring Inerio	
	MVR12		+ 1.2V to 37V	350p	SPST on off	55p
	1A -5V		LM317H 5W 1-5A		SPDT C over	65p
25p	1A -12V	2200	+ 1.2V to 37V	320m	DPDT 6 Tag	85p

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LES HOLDER Dome shaped, Red, Blue	
Green, Yellow, White 16	
LES BULBS 6v and 12v 9;	
MES HOLDERS Chrome cover, Red o	
Amber, Jewelled top 50;	
LES OR MES Batten Holders 16	ø
MES BULBS 3-5V 6V 12V 9	
<b>NEONS</b> Mains, Sealed with Resistor	
Sq. Top, Red or Grn. Round Top Red 24s	9
Neon with leads, 95V AC (No resistor	)
91	0

TRIACS 3A400V 6A400V 8A400V 10A500V

9p 17p

11b bag	Anhydrou	s 65p + 3	0p p. &
DALO I	Spare tip	SIST	75
COPPE	RCLAD	BOARDS'	
Fibre	Single-	Double-	SRBP
Glass	sided	sided	7" x 8"
6" x 6"	75p	90p	60 p
6" x 12"	130p	175p	

6" x 12"	″ 130p	175p
CASE NJHC NJSF 2 VERO 75 1239 1237J 1238D 1798K	1 85p 2 275p	We stock many more cases, Vero, Classic NJHC, NJSF ranges Classic, our Vinyl- coated low cost attrac- tive steel cases (P. Range).

	-
SWITCHES* PUSH BUTTON	:
Push to Make 15p Push to Bre	ak <b>2</b> 5p
SP changeover centre off	28p
ROCKER: (black) on/off 10A 250	V 23p
ROCKER: Illuminated (white)	
lights when on, 3A 240V	52p
ROTARY: (ADJUSTABLE ST	OP) 1
pole/2-12 way, 2p/2-6W, 3p/2-4W	/. 4p/2-
3W 41p ROTARY: Mains 250V AC, 4 Am	p 42p

0	ROTART: Mains 250V AC, 4 Amp	42P
- 1	KNOBS fit i" shaft with grub scre	aws
	except K2 (push fit) & K8 (for sliders	).
:	KI Black or White pointer type	9n
1	K4 Black serrated. Metal top with lin	e
9	indicator 33mm dlam.	22p
	K4A As above but 25mm dlam.	28p
	K5 Black fluted metal top and skirt	
п	calibrated 0-10, 37mm dlam.	26 p
	K6 PK2 as K5, pointer on skirt	26p
	K7 Black, knurled, tanered, Metal	
	top & skirt, Calib. 0-10, 30mm	25p
•	K8 Black or silvered for slider not	160
	K9 Solld Aluminium, Amplifler Kno	b.
e	Professional type, with etch line	
,	indicator 16.5 × 12mm diam.	26p
,	K10 As above but tapered 18.5 × 17	mm
- 1	diam.	28p
-	K11 Aluminium, (Top Hat) Knurled	
J	body with 18 5mm skirt, Knotche	d
	Attractive	30p

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#### ALL PRICES INCLUDE VAT

#### **NEW EDU-KIT MAJOR**

Completely Solderless Electronic Construction Kit. Build these projects without soldering iron or solder



★ 4 Transistor Earpiece
Radio
★ Signal Tracer
★ Signal Injector
★ Transistor Tester
NPN-PNP

4 Transistor Push Pull

Amplifier

5 Transistor Push Pull

★ Electronic Metronome

★ Electronic Noise \*

Transistor Push Pull

\*\*Electronic Noise Generator\*\* Amplifier

★ 7 Transistor Loudspeaker Radio MW/LW

★ 5 Transistor Short Wave Radio

\* Batteryless Crystal

Radio
One Transistor Radio
2 Transistor Regenerative Radio

tive Radio

★ 3 Transistor Regenerative Radio

★ Audible Continuity
Tester

★ Sensitive Pre-Amplifier

Components include: 24 Resistors 

\$\pm\$ 21 Capacitors 

\$\pm\$ 10 Transistors 

\$\pm\$ 5" \times 3" Loudspeaker 

\$\pm\$ Earpiece 

\$\pm\$ Mica Baseboard 

\$\pm\$ 3 12-way connectors 

\$\pm\$ 2 Volume controls 

\$\pm\$ 2 Slider Switches 

\$\pm\$ 1 Tuning Condenser 

\$\pm\$ 3 Knobs 

\$\pm\$ Ready Wound MW/LW/SW Coils 

\$\pm\$ Ferrite Rod 

\$\pm\$ 6\frac{1}{2} yards of wire 

\$\pm\$ 1 Yard of sleeving, etc. Complete kit of parts including construction plans.

VHF AIR **CONVERTER KIT** 

Build this Converter Kit and receive the Aircraft Band by placing it by the side of a radio tuned to Medium Wave or the Long Wave Band and operating as shown in the instructions supplied free with all parts. Uses a retractable chrome plated telescopic chrome plated telescopic aerial, Gain Control, V.H.F. Tuning Capacitor, Transistor,

All Parts including Case and

P & P + Ins. £1.10

£3.95 Plans. P & P and Ins. 60p.

#### POCKET FIVE

NOW WITH 3 LOUDSPEAKER

LOUDSPEAKER

3 Tuneable wavebands.

M.W., L.W., and Trawler

Band, 7 stages, 5 transistors and

2 diodes, supersensitive ferrite rod
aerial, attractive black and gold case,
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All Parts including Case and Plans.

Total Build-ing Costs

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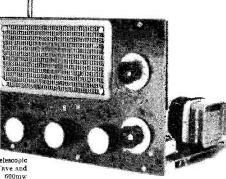
MULTIBAND V.H.F. AND A.M. RECEIVER 13 TRANSISTORS AND FIVE DIODES QUALITY 5" × 3" LOUDSPEAKER

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Aerial, angled and rotatable for peak Short Wave and V.H.F. reception. Push Pull output using 600mw Transistors. Gain, Wave-Change and Tone Controls. Plus two Slider Switches. Negative Feedback circuit and SPECIAL POWER BOOSTER SOCKET AND

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E.V.5. 5 Transistors and 2 diodes. MW/LW. Powered by 44 volt Battery. Ferrite rod areial, tuning condenser, volume control, and now with 3" loudspeaker. Attractive case with red speaker grille. Size  $9'' \times 5_2^* \times 2_1^{*'}$  approx. All parts including Case and Plans.

Total Building costs £4.30 P&P + Ins. 80p.

E.V. 6. Case and looks as above. 6 Transistors and 3 diodes. Powered by 9 volt battery. Ferrite rod aerial, 3" loudspeaker, etc., MW/LW coverage. Push Pull

All parts including Case and Plans.

Total Building costs £4.95 P&P + Ins. 90p.

E.V. 7. Case and looks as above, 7 Transistors and 3 diodes. Six wavebands, MW/LW, Trawler Band, SWI, SW2, SW3, powered by 9 voit battery. Push Pull output. Telescopic aerial for short waves. 3" All parts including Case and Plans.

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#### CONSTRU

ECK 2 Self Contained Multi-Band V.H.F. Receiver Kit. 8 Transistors and 9 Diodes Push/Pull output. 3" Loudspeaker. Gain Control.7 section chrome plated telescapic aerful, V.H.F. Tuning Capacitor, Resistors, Capacitors, Transistors, etc. Will receive T.V. Sound, Public Service Band, Alricatit, V.H.F. Local Stations, etc. Operates from a 9 Vult PP. Battery (not supplied with kit). Complete kit of parts including construction plans.

£7.95 P & P and Ins. 90p.

FCK4 7 Transistors, 6 tuneable wavebands, MW. LW, Travler Band, 3 Short Wave Bands. Receiver Kit. With 5" > 3" Loudspeaker. Push/Pull output stage, Gain Control, and Rotary Switch. 7 Transistors and 4 Diodes. 6 section chrome-plated telescopic aerial. 8" Sensitive Ready Wound Ferrite Rod Aerial. Tuning Capacitor. Resistors, Capacitors, etc. Operates from a 9 Volk PPJ Battery (not supplied with kit). Complete kit of parts-including construction plans. £7.25

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ects without Soldering Iron or Solder.

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\*\* One Transistor Regenerative Radio.

\*\*2 Transistor Regenerative Radio Medium Wave Coverage.

\*\*Radio Medium Wave Coverage.

\*\*Coverage No Medium Wave Coverage.

\*\*Coverage No Medium Wave Coverage.

\*\*Complete kit of parts including Loudspan State Complete Radio Aerial, Capacitors, Resistors, etc.

Radio Medium Wave
Coverage

 ★ 4 Transistor Medium Wave
Loudspeaker Radio

 ★ Electronic Noise Generator

 ★ Electronic Noise Generator

All parts including Loud-speaker, Earplece, MW Ferrite Rod Aerial, Capacitors, Re-sistors, Transistors, etc. Complete kit of parts includ-ing construction plans.

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