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Published by IPC Magazines Ltd, Fleetway House, FarrIngdon Street, London EC4A 4AD. Tel. 01-634 4444

## BINDERS

Blnders £2 10 and indexes 45 p (inc. $\vee A T$ ) can be supplied by the Post Sales Department, IPC Magazines Ltd., Lavington House, 25 Lavington Street, London, SE1 OPF.
Remittances with overseas orders for binders should include 60 p to cover despatch and postage.
Due to an eniarged format, shortly to be announced, there will be a delay in supplying larger sized binders for Volume 53. See page 279.

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# Morse Code~ the PW Way 

AMONG the offerings in this issue of Practical Wireless is a Morse Code Tutor, an excellent project that we are sure will be suitable for construction by amateur radio clubs and groups up and down the country as well as by many individual constructors. While clubs often have organised code practice classes under the eagle eye of a qualified instructor it is very convenient if the odd member or two can sit down with our Tutor and get it to send perfect code over a wide range of speeds, just for practice.

Whether we like it or not, passing the Morse proficiency test as set by the Home Office is a necessary requirement before the transmitting amateur can reach out on the HF bands. There is no other way out.

Many Class "B" licencees, restricted to 144 MHz and above, are quite happy to leave it at that. However, there generally comes a time when a two-way link on one of the HF bands would be most useful in conjunction with some VHF activity or other. The Oscar amateur radio satellite programme is one case in point where a world-wide network is maintained by enthusiasts on the HF bands, passing a continuous stream of scientific information concerning the operation of the satellites and predictions for future orbits. Without that network the full potential of these satellites would never be realised.

While it is very easy to spurn the Morse code and to devise various reasons why one should not learn it, the fact remains that those who do take the trouble to pass the test are frequently delighted to discover that they have, without a great deal of trouble, acquired the ability to converse in a new language! However, that is only the beginning and after a few months of experience on the key the newly-acquired art of communication can easily supercede the rather artless operation of just talking into a microphone! If $D X$ is the main interest it will be found that many countries around the world appear at the low frequency end of bands on CW, some of which seldom appear on telephony.
Unfortunately, old-fashioned methods of teaching the code are still persisted in, such as working up very slowly from a few characters a minute to 12 words a minute or so but always religiously maintaining the correct ratio of dots and dashes. This leads, almost inevitably, to the development in the pupil's mind of the well-known "speed barrier" around 7 or 8wpm when any further progress seems impossible regardless of the amount of practice indulged in. The only sensible method is to learn the sound of the whole letter or number sent at fairly high speed but with long gaps in between. Note that the sound is learned rather than the individual dots and dashes. Now when the gaps are gradually reduced in length, suddenly, hey presto, code at 12wpm without any "barrier"!

The PW Morse Code Tutor can do just that without the need for an experienced operator to be present at the time. Any letter or character can be repeated ad lib until it is known by heart or rather, by ear! Mix them up and send them at random thus avoiding any chance of self-deception which always occurs if the letters etc are sent in alphabetical order as is sometimes advocated.
Once confidence has been gained there is no reason why further practice should not be obtained with the help of another experienced operator but avoid other inexperienced operators like the plague! Otherwise you will both finish up with the same faults and these can be very difficult to eradicate.

Eric Dowdeswell Assistant Editor

| Tilonel Iowes |
| :---: |
| T E are sorry to have to ${ }^{\star}$ report that Lionel |
| Howes G3AYA has |
| * to relinquish the post of Editor* |
| * because of continuing |
| $\star$ health. Lionel was previously ${ }^{\text {® }}$ |
| $\star$ Assistant Editor before being |
| $\star$ appointed Editor in Octobe |
| غ 1974. |
| We are sure that reade |
| many friends of Lio |
| ${ }_{\star}$ want to wish him a very spe |
| $\star$ recovery to good health |
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| ${ }_{\text {t }}$ tinued involvement wi |
| $\star$ magazine in the future. |
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W
E have been asked by $A$. Marshall (London) Ltd. to inform readers that this is their correct name, although from time to time a different name appears in the index to advertisers. The correct address for this company is 42 Cricklewood Broadway, London NW2 3ET. Tel: 01-452 0161.

## Ten more to remember!

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The final two in this series are the BC549/550P and the BC559/ 560 P and these are intended for low noise input stages in tape recorders and other good quality audio equipment.

Ferranti Ltd., Special Components Division, Gem Mill, Chadderton, Oldham. Tel.: 061-624 0515.

## Weekend report

EIGHTEEN of their 25 members attended a weekend meeting of the St. Dunstan's Amateur Radio Society held in the Winter Garden at Ian Fraser House, Ovingdean, near Brighton, on May 20-22. Among those who had travelled some distance to be there were their Secretary/Treasurer Ted John G3SEJ, Merseyside, Jock Inness G4AJP, Yorkshire and SWL Tony Parkinson, Lancashire. The Society, open to St. Dunstaners, was inaugurated in January 1976 and is affiliated to the RAIBC, and RSGB, because of their military associations, to the RAFARS, RNARS and the Royal Signals ARS.

While using the call sign G4AFV/A, instead of their regular call GB3STD for such meetings, they worked several stations in the USA on 20 m . Using a KW Atlanta feeding a Moseley Beam they also had a QSO with JA2GPR, despite heavy QRM. The operating of Bill Shea G4AUT impressed the visitors among whom were Louis Varney G5RV President of the Mid-Sussex ARS, Alan Baker G8LGQ Chairman of the Brighton And District Radio Society, and a party of cadets from the 2464 (Storrington) Squadron of the ATC. At the same time a 2 m transceiver was being operated by Peter Bargen G30TB, from Aylesbury.

The Saturday afternoon lecture was given by Ron Ham on 'Satellite Communication' and this was recorded by Ralph Cathles G3NDF for other members of the 'white stick' brigade. After the
lecture Joan Ham showed the St. Dunstaners a selection of wartime morse keys including the Muirhead key made in 1915 and used lby the late Nelly Corry G2YL when she worked all continents on 10 m in October 1935. George Cole G4AWI a keen CW operator referred to the keys and to the value of CW operating when he moved the vote of thanks to the speaker. Looking to the future, Secretary Ted John said that it is proposed to set up a permanent amateur radio station at Ian Fraser House and to have it working by the end of the year.

## Seaside amateurs

SOME 30 local radio enthusiasts were present at the inaugural meeting of the newly formed Brighton and District Radio Society, held at the St. John The Baptist Social Club Hall, Bristol Road, Brighton, on May 18th. The chairman, Alan Baker, G8LGQ, said that their ainns are to further the well-being of amateur radio in all it's aspects, provide a technical advice service, help members with home construction, and promote good relations between people of similar interests.

They meet fortnightly at the Baptist Hall, where new members are welcome. Details of membership and their future programmés can be obtained from the secretary, Nigel Hewitt, G8JFT, 74 Carlyle Street, Brighton.

## Welsh date

NOW established as the major amateur radio event in Wales the Welsh Amateur Radio Convention will again be held this year, on Sunday, 25th September, at the usual venueOakdale Community College, Oakdale, Blackwood, Gwent. The Convention will be opened by the Mayor of Islwyn Borough.

Apart from the usual Trade Exhibition, the programme will include an exhibition of early amateur radio equipment, an
exhibition station, illustrated lectures on a visit to the Seychelles Islands by Roger Brown, G3LQP/ VQ9RB, and the launching and post-launch control systems for AERIEL 5, one of the U.K. satellites, by John Wright, G3VPW, of the Appleton Research Labora--lory, Slough.

There will be increased parking space this year and advice on obtaining overnight accommodation will be available upon request, from R. B. Davies, GW3KYA, 16 Vancouver Drive, Penmain, Blackwood, Gwent.

## New WRC times

Afrom Wednesday, 7th September, the transmissions of World Radio Club will be as follows:-
Wednesday - 0815 to 0830 GMT Wednesday - 1330 to 1345 GMT Wednesday - 2315 to 2330 GMT Friday - 2100 to 2115 GMT As can be seen the 0815 Sunday transmission has been cancelled.

## University convention

THE Lancaster University Amateur Radio Society North West Amateur Radio Convention (must be the longest title of a convention yet) will be opening their doors to the general public at 2 p.m. on Saturday, 17th September, and again on Sunday 18th.

Lectures will cover many topics 'including repeaters, Oscar, TV, aerials, QRP, etc. There will be a supporting programme of films a raffle, constructors competition, and a coach tour of the lakes on Sunday. Full accommodation and all meals, including a dinner on Saturday are available. Optional lunch on Saturday will be available for those arriving early.

Further information concerning the convention and charges for the weekend can be obtained from J. R. Morris, Dept. of Physics, Lancaster University, Lancaster.

## Ma.nd 7~Minio  I. HICKMAN

THIS power supply was designed to be easy to construct, with a very simple circuit, yet provide facilities and a standard of performance normally only associated with professional equipment. It is constructed in a diecast box which also acts as heat sink for the power transistor. The dimensions of the box used limit the size of mains transformer but this unit provides 0 to 12 V at up to 500 mA . However, with a larger box and heat sink, the same circuit can provide 1A with very minor modifications.

## CIRCUIT

The circuit diagram is given in Fig. 1. Care has been taken to ensure that the circuit is stable under all conditions of use and some of the finer design points are explained below, along with the circuit description. The centre-tapped transformer and bridge rectifier provide both positive and negative rails, each of 15 V nominal ( 18 V off load). The negative rail provides the main current, the positive


rail being used as an auxiliary supply for the reference zener D1. C2, C3 and the zener provide three stages of smoothing/filtering, reducing hum across D1 to a fraction of a millivolt. VR2 allows the range covered by the output voltage control VR3 to be set to 12 V . The stabilised output voltage is less than the voltage at the slider of VR3 by the $\mathrm{V}_{\mathrm{BE}}$ of Tr2, about 600 mV . Tr2, 3 and 4 form a triple compound emitter follower, the large internal loop gain

Fig. 1: Complete circuit diagram of the Handy-Mini Power Supply.
The lettered points wIII assist in connecting the leads from the board to the components mounted in the other part of the box.


Fig. 2: above, gives details of the holes required in the lid of the box. The hole for the meter may need to be altered to suit the meter actually used.
Fig. 3: below, shows the wiring between components and the components layout. In many cases the wire ends of components are long enough to complete the wiring. Note that resistance R3 is wound in the form of a coil, for convenience only.

## components list



ensuring an output impedance of $0.1 \Omega$ typically.
A separate output switch $S 2$ ensures that the voltage across the load collapses to zero in a second or so after switch-off. If, as with some power supplies, a mains switch only is provided, some output voltage can still be present seconds after switch-off. This can result in damage to circuitry fed from the power supply during subsequent soldering operations. C5 keeps the output impedance low at high frequencies. D2 catches voltage reversals on switch-off when feeding an inductive load and also prevents damage to the power supply should it be connected in series with another supply of higher current rating. D2 should therfore be rated at 1A, at least.

Trl protects the power supply in the event of an overload at the output, by reducing the output voltage. Should more than 50 mA be drawn through R 4 , or more than 500 mA through $\mathbb{R} 4$ and $R 3$ in parallel, Trl turns on and reduces the reference voltage by shunting D1. The output voltage therefore falls just as far as is necessary to prevent the current exceeding 500 mA on the high current range or 50 mA on the low current range. The low current
range is very useful when testing out a variety of low power circuits, since 50 mA for a few seconds does not in practice damage any common semiconductor. Circuits can thus be switched on in safety without bothering to check them through too thoroughly! Meter M1 measures the voltage drop across the current sensing resistors R3 and R4 and thus indicates the current drawn by the load connected to the supply, except for the small amount of current through R8.

C4 starts to roll off the loop gain at around 100 Hz . In normal operation (constant voltage mode) a much smaller capacitor could be used, as further gain roll off is provided by $C 5$. However, in the constantcurrent mode with a short-circuit at the output terminals, C5 obviously can be ignored and there is the additional loop gain provided by Trl to cope with as well. Thus, it is the current limit condition which determines the loop shaping capacitor, but it has been possible to retain the full loop gain up to 100 Hz and this together with the low ripple across D1 results in the output ripple at full load being only $200 \mu \mathrm{~V}$ RMS.


Fig. 4: Wiring of the mains supply components and Trf in the box. This may be compared with the photograph below. The leads of Tr4 are cut short and wire leads soldered on to reach to the board, Ensure that these loads are completely insulated to avoíd contact with the case. Note that an insulation kit must be used with this transistor.


A view of the finished power supply. Note that capacilor C5 is mounted directly across the output terminals, observing the correct polarity of this electrolytic.
Front panel drillings are given in Fig. 2. The meter is the popular type MR38P and serves as one support for the circuit board, the other end being supported on a piece of studding, glued to the back of the front panel. The author made up the circuit on a piece of $0 \cdot 1$ in pitch plain pin-board. Component leads were passed through holes and wired up on the back of the board. Crossovers were avoided so that the same layout can be used for a
printed circuit. See Fig. 3 which shows the component side of the board with connections ghosted in. Fig. 4 shows the wiring in the box itself.

The following components mount in the box as shown in the photograph; fuse F1, mains switch S3, mains transformer Tl, bridge rectifier Bl, capacitors Cl and C 2 and Tr 4 . The latter is mounted inside the box, with a mica washer insulating set and the base and collector pins are cropped fairly short, with the leads from them covered by an insulating channel.

Note that the transformer is mounted so as to fit between S1 and S2 when the lid is fitted.

## CALIBRATION

When construction is complete, set VR1 and VR2 to mid travel and switch on. Check that Output Voltage control VR3 varies the output voltage and adjust VR2 so that with VR3 fully clockwise, the output voltage is 12 V . Mark in calibrations 0 to 12 V round the Output Voltage control VR3. Set Current Range Switch Sl to 50 mA (up) and connect an AVO on its 100 mA range across the output terminals. Adjust VR1 for 45 mA short circuit current. Adjust VR4 for full scale reading on $500 \mu \mathrm{~A}$ meter M1. With Current Range Switch S 1 set to 500 mA (down) check that M1 reads the same as the AVO, i.e. approx. 500 mA . If not, adjust R 3 which can conveniently be 5 in of 36 SWG Eureka wire. Finally, re-adjust VR1 if necessary to set the short circuit current to exactly 500 mA . This is important to avoid overrunning the transformer. Note that on the 50 mA range, the short circuit current in the load will actually only be 45 mA , the other 5 mA represents the current through D1.

## USES

In use, this power supply is practically indestructible. It may be connected in series with another supply to give a higher voltage but should not be connected in parallel with a supply which has a higher maximum output voltage. It may be connected in parallel with another 0 to 12 V supply to give a higher current. The mains switch S3 should be left on and the output turned on and off with output switch S2.


THIS unit is intended as a self-teaching aid for learning Morse by presenting the code as a series of sounds rather than dots and dashes.
The system uses a set of coded cards, each representing a Morse character which may be a letter, figure or standard phrase. When a character card is plugged into the unit, the appropriate code is reproduced as an audible and/or visual signal. A manual control alters the speed of the code as the user becomes more proficient at recognising the character; whatever speed is chosen, the relative spacing between elements of the character is maintained (see Fig. 6). This ensures that the user develops accurate timing. Controls are also provided to adjust the volume and tone of the Morse to the most natural sound

The unit can also be used as a Morse practice oscillator when required, either for normal sending practice or to copy signals just produced by the Morse generator.

## Principle of operation

The system uses a seven bit binary code; the most significant bit(s) represent a binary equivalent of the Morse character where logic 1 is the equivalent of a 'dash', and logic 0 is a 'dot'. The least significant bit(s) identify the beginning of the character. Each bit is sampled in turn from right to left. If the first bit is a logic 0 , it is ignored as are successive logic 0 s . Once a logic 1 is encountered, this informs the circuit that all the remaining bits are binary Morse code.


The code cards slot into the front panel edge connector. Pressing the start button initiates the reading process.

Fig. 1 shows the seven bit binary code representing the letter $Y$. In this case, the first two bits are sampled, seen to be logic $0 s$ and ignored. The next bit is also ignored, but as it is a logic 1, the system prepares to reproduce the remaining bits as Morse code elements. Thus the next logic 1 is produced as a 'dash', the logic 0 as a 'dot', and so on until all seven bits have been sampled.


Fig. 2: Construction of the code cards. The artwork for the cards is shown on page 267. Alternatively, ready made cards may be obtained through the Readers' PCB service. Cards may be constructed from hard wired, suitably cut, verostrip. 0.7" matrix board should be used.

## $110110010 \leftarrow$ Binary code

Since Morse characters contain up to six elements (except in one case which has been ignored to avoid unnecessary complication) then a maximum of seven bits is needed for binary encoding. Where a character has less bits, these are made up to seven by the inclusion of logic 0 s as seen in the example above. In this way, the end of a code is easily recognised by preventing further operation of the system once seven bits have been sampled.


[^1]
## Code cards

The seven bit code is applied to the unit by using a patch card of copper laminate board etched to represent the code to be reproduced. Fig. 2 shows how the patch card, representing the letter ' $Y$ ', is used to apply the correct code to the data select inputs of IC3. With no card in position, inputs 1 to 7 of IC3 are held at logic 1 by 'pull-up' resistors. It can be seen that if the card is plugged into the connector, connection A applies 0 V (i.e. logic 0 ) to the copper of the patch card, which in turn 'pulls' down inputs 1,2 and 5 of IC3 to logic 0 . Inputs 3 , 4,6 and 7 remain held at logic 1 .

Since no code requires more than seven bits, input 0 is permanently wired to logic 0 , obviating the need for this to be supplied by the patch card.

## Circuit description

Looking at Fig. 3, the circuit comprises an astable clock oscillator (ICl) a four bit binary counter (IC2) a one of eight line data selector (IC3) controlled by the binary counter, a pair of J-K flip-flops, an audio oscillator and a series of gates.

A character card inserted in the edge connector provides a binary pattern equivalent to the Morse character on the data input lines of IC3. On pressing S3, the circuit clock ICl increments the binary counter causing the data bits on the character card to be interrogated in turn. Ignoring all leading 0 s , the circuit scans until it encounters a 1 . This sets up the bistables which will change state in a manner dependent on subsequent high/low states which make up the character card. If the next bit is a 0 , then flip-flop IC4a changes state and gates the audio

oscillator for one clock cycle. If a 1 is encountered, the flip-flops act as a divide by 3 routine for a duration of three clock pulses; similarly, this gates the audio oscillator on for a period of three clock pulses. Since reset action requires a duration of one clock pulse, the 1:1:3 dot-space-dash relationship is absolutely textbook and immutable. Further replays of the Morse character can be made by depression of S3 to enable the master reset line.
VR1 and VR2 provide control of the clock and audio oscillator respectively; the first controls the speed of character transmission and the second adjustment changes the tone. IC5 provides enough signal to drive a high impedance loudspeaker directly. The prototype incorporates an LED which


Fig. 3: Complete circuit diagram of the Morse Tutor. The switches are shown set to 'Read' mode and 'AudiolVisual', Pins 1 and 10 of the edge connector are not shown since these are not required for circuit operation. C7, not shown, is connected across Vcc and OV.
is also driven by the oscillator to give a visual Morse output. However, this is really just a gimmick. The prototype was powered by a 6 V battery. Since the logic ICs require 5 V , silicon diode D 2 drops the positive rail voltage by about $0 \cdot 7 \mathrm{~V}$.


## Construction

The prototype was constructed in a diecast box drilled as in Fig. 4. If alternative components are used, care should be taken to ensure that they do not foul each other or the logic board before the mounting holes are drilled as there is not a great deal of space available in the size of box used.

The edge connector should be cut down in size so as to leave twelve contacts. The oontacts at each end should then be removed to leave room for insertion of the mounting feet. This assembly is mounted through the panel aperture from the underside with its contacts nearest to the panel edge. The 'least significant bit' end of the connector will then be towards the centre of the panel. After wiring, the speaker bracket may be attached to the connector mounting screws using stand-off pillars. The speaker can be attached to this by means of its terminals.

Fig. 7 shows the board used to form 48 individual patch cards. The shaded areas represent copper, the remainder being etched away. These cards have been designed to fit in a ten-way edge connector with closed ends, and should be trimmed carefully for accurate positioning. If another make of edge connector is used, some modification to the cards may


All dimensions in mm
All holes 3 mm . dia
W491
Fig. 4: Drilling details.
be necessary. Additional cards may be made in a similar fashion for those characters that have been omitted. If preferred, they may be made up from small pieces of $0 \cdot$ lin matrix veroboard by connecting the appropriate copper strips together with wire links.


Close up view of edge connector wiring and earpiece mounting bracket. Note detalls of edge connector truncation.

## components list



Board A


Fig. 5: Veroboard layout. Board A mounting holes are 3 mm dia with 63 mm spacing. They attach the board to diecast box using the rubber feet mounting screws with spacers.
Board $B$ attaches to diecast box using the 6 V power termina/s. Board C is attached to the Key terminals. The flying leads shown on both boards connect with the appropriate terminals.

## PIN CONNECTIONS

| A Edge Connector pin 3 |  |  |  |  | S1a (read) |
| :---: | :---: | :---: | :---: | :---: | :---: |
| B | " | " | , |  | VR1 |
| C | " | " | " |  | VR1 |
| D | " | " | " |  | VR2 |
| E | " | " | " |  | VR2 |
| F | " | " | " | R | OV terminal |
| G | " | " | " |  | Sib (Y Fig. 3) |
| H S3 |  |  |  | T | Board A pin L |
|  | Aux | put |  |  | $V$ and S1b |
| K | S1a |  |  |  | S1a (Key) |
|  | Boar | pin |  |  | Board B pin T |

Resistors
R1 to $94 \cdot 7 \mathrm{k} \Omega$
R10 $\quad 10 \mathrm{k} \Omega$
R11 $\quad 220 \mathrm{k} \Omega$
R12 $10 \mathrm{k} \Omega$
R13 $\quad 1 \mathrm{M} \Omega$
R14 $\quad 180 \Omega$
R15 $470 \mathrm{k} \Omega$
All $4 W 5 \%$

Semiconductors

| Tr1 | BC108 |
| :--- | :--- |
| IC1 | 555 (8 pin) |
| IC2 | 7493 |
| IC3 | 74151 |
| IC4 | 7473 |
| IC5 | $555(8 \mathrm{pin})$ |
| IC6 | 7404 |
| IC7 | 7400 |
| D1, D2 BAX16 |  |
| D3 | LED |

## Capacitors

C1 $0.22 \mu \mathrm{~F}$ Polyester
C2 $0.01 \mu \mathrm{~F}$ Polyester
C3 820pF Polystyrene
C4 $0 \cdot 1 / \mu \mathrm{F}$ Polyester
C5 $100 \mu \mathrm{~F} 10 \mathrm{~V}$
C6 $470 \mu \mathrm{~F} 10 \mathrm{~V}$
C70.1ıF Disc ceramic

## Switches

S1 DPDT Centre off miniature toggle
S2 DPDT
S3 SPDT push to make

## Potentiometers

VR1 $1 \mathrm{M} \Omega$ Carbon linear $\frac{1}{2} \mathrm{~W}$.
VR2 1 MS Carbon linear $\frac{1}{2}$ W.
VR3 $100 \Omega$ Wirewound 1 W .

## Miscellaneous

16 Way $0.1^{\prime \prime}$ edge connector, 3.5 mm earphone socket, IC sockets, 2 mm terminals, rubber feet, diecast box $114 \times 89 \times 55 \mathrm{~mm}$, knobs, nuts and bolts, 6 mm mounting pillars, veropins, veroboard, battery 6 V , LS1 PO type receiver insert (30 ( ) .
Note that the edge connector and most other components are RS types available from Doram and other suppliers.

## Testing

A 6 V battery or suitable power supply should be connected to the unit. Turn SPEED control VR1 to minimum and VOLUME control VR3 to maximum. Set S2 to '\&' position (i.e. VISUAL \& AUDIO). If the START button is pressed and released, six long bursts of tone should be heard, along with a visual signal. This corresponds to six 'dashes' and is the signal produced when no patch card is in position.

A spare piece of unetched copper laminate board cut to the size of a patch card should now be inserted in the connector. Operation and release of S3 should not result in any output signal (allowing time for all inputs to be read) as no logic 1 will be encountered. These procedures ensure that the unit is sensing the condition on each edge connector input pin correctly

Finally, the card representing 'underline' is recommended for use as a test card as it checks the unit

for correct response to changes from dots to dashes and vice versa with a sequence of two dots and two dashes.

Check the remaining two positions of $S 2$ for the appropriate output signal, and that the volume, tone, and speed controls are effective. It is wise to ensure that each patch card produces the desired oode, and that the copper accurately covers the appropriate edge connector pins without any chance of contact with pins that should remain unconnected.

Whenever power is first applied to the unit, it is possible that an output signal will result. This can also occur if Sla is switched from KEY to READ. This can be cleared, either by holding down $S 3$ until a card is inserted, or releasing it and waiting for the counter to complete a cycle. It was felt that this problem did not really justify the extra complication of some form of power-on reset.
The cards can be selected either according to some set learning system or at random and the character identified from the face of the card. Constant replay of the pattern become associated with the letter. The object of the learning exercise should always be to familiarise Morse characters in terms of a sound pattern rather than as discrete dots and dashes.
It is recommended that short groups of letters are familiarised at a time-perhaps about five-until the whole alphabet can be recognised from the dah-di-dah rhythm of the characters. Although the unit may be used as a practice oscillator, the beginner's studies should be concentrated on receiving Morse


Preliminary call Full stop Comma Apostrophe Hyphen Inverted commas Question Oblique stroke Brackets Underline Long break sign Error End of message
End of work

Spacing and length of signal
(i). A dash equals 3 dots
(ii) The space between elements which form a character is equal to 1 dot (iii) The space be tween 2 characters equals 3 dots
(iv) The space between 2 words is equal to 5 to 7 dots

Fig. 6: The Morse code.
rather than sending. Once the former has been attained, the latter is easy. Although the Morse Tutor is very useful, it provides no substitute to reading Morse code from an experienced operator since sequential groups of letters cannot be generatedindividual character recognition of the alphabet won't ensure a pass in the Morse test! However, using the unit in conjunction with records, tapes and RSGB Slow Morse transmissions in particular should result in a pass.


## Television

## - INTRODUCING THE PHILIPS G11 CHASSIS

The first entirely new colour chassis from Philips since 1970 is a major event. This new chassis, designed around the 20AX tube, is intended to serve Philips well into the 80 s . Great attention therefore has been paid to getting things right. The first of a two part article on the development of this chassis and the new techniques it uses

## - JAPANESE COLOUR SET FAULTS

Despite their reputation for fault-free service, Japanese colour sets must inevitably eventually fail as the effects of age appear. Peter Murchison reports on recent experiences with a selection of sets to show the sorts of troubles you can expect

## THE TV LOGIC STATE CHECKER

The increasing use of digital i.c.s in TV circuitry and allied fields (e.g. our Teletext decoder) means that special fault-tracing techniques are required when problems develop. The construction of a simple in strument which indicates the "logical state" of every pin of a digital i.c. will be described. The device derives its power from the i.c. under test and uses l.e.d.s as indicators. It's housed in a small diecast box with an external i.c. clip to hook up to the i.c. being investigated Very useful for anyone dealing with logic circuits

## TELEPRO COLOUR RECEIVERS

The Telepro range of colour receivers was made by a Telefusion subsidiary and used a chassis based on the Decca 30 series chassis. The sets have their own fault patterns however, and Barry Pamplin provides servicing hints on dealing with them

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## Fuńction

The front end accepts a signal from the aerial at a frequency in the range 88 to 100 MHz . The signal is amplified and passed through a number of tuned circuits which attenuate unwanted signals. The amplified signal is then passed to the mixer stage where it is mixed with a locally generated signal; the latter is always kept 10.7 MHz away from the signal frequency being received by the aerial.

The output from the mixer stage is a 10.7 MHz difference signal frequency modulated with the audio signal. This is fed through a single or double tuned resonant circuit before being passed to the output part of the front end from where it is processed by the limiter/amplifier/demodulator.

## Ganged capacitor types

Some types of FM front ends employ conventional ganged capacitor tuning. The low capacity VHF ganged capacitor usually has another variable capacitor mounted on the same spindle for AM medium and long wave tuning. Such front ends incorporating an AM tuning section are convenient for obtaining FM and AM reception using a single tuning knob although there is no AM circuitry other than the variable capacitor itself in the front end.
Ganged capacitor tuned front ends are available with a spring loaded reduction drive fitted to the spindle. The tuning knob may be fitted directly to such a front end, but provision must be made for fitting a tuning scale and pointer. Some other types do not have a reduction drive and some form of slow motion drive must be designed; for example, a suitable arrangement of drive cords may be used to provide the required speed reduction.

## Varicap tuning

This is the other tuning system widely used in modern front ends. Such front ends require a tuning voltage which may be obtained from a potentiometer. One of the advantages of using a varicap system is that the unit may be placed at any convenient point in a receiver, even if the tuning control must be placed at the opposite side of the cabinet. Capacitor tuned front end placement is


usually governed by the requirement that the tuning spindle must be at a suitable place for the drive cords or tuning knobs to operate it.

High performance front ends require quite a number of signal tuned circuits, but the number of variable capacitor sections which it is convenient to fit on a single ganged capacitor spindle is limited to about 3 or 4 . There is no such limitation with varicap diode tuning and therefore the most expensive, high performance front ends tend to employ this system.

Another advantage of varicap tuning is that it is quite easy to arrange for the switched tuning of a number of selected frequencies; this is very desirable in domestic receivers. The varicap system requires a regulated supply from which the tuning voltage is derived to prevent drift. Further, the regulator circuit must reduce the power frequency hum level to a very low value on the varicap tuning supply line; any residual hum manifests itself as hum in the audio output. A hum level in the order of 1 mV or less is required.

## Typical circuit

A typical varicap tuning circuit is shown in Fig. 1. The unregulated input voltage from the receiver supply may be fed to the input of a low output current voltage regulator integrated circuit (such as the $\mu$ A78L15 or the TBA625C for a 15 V output or the ${ }_{\mu} \mathrm{A} 78 \mathrm{~L} 12$ or the TBA625B for a 12 V output). While discrete regulator circuits can be constructed, the use of a monolithic regulator simplifies the circuitry.

The hum level at the output of the regulator should be some 50 to 60 dB lower than that at the input of the regulator. It is important that the reservoir capacitor Cl should have an adequate value or the hum level at the input to the regulator will be high leaving an undue level on the output. C2 is a small non-electrolytic capacitor in parallel with Cl to provide high frequency decoupling.

The stabilised output voltage from the regulator is applied to a number of potentiometers in series with the resistor R1; this resistor ensures that the tuning voltage never falls below about 2 to 3 V since the front end unit does not operate at very small tuning voltages.

The potentiometer in use is selected by S1. The tuning potentiometer VR1, selected by S1, enables manual band coverage. The other positions of S1 are for the reception of switch tuned frequencies. VR2 to VR6 inclusive should be high quality trimmer potentiometers with about 15 or 22 turns each for ease of adjustment; single turn potentiometers are unsuitable. Similarly, VR1 may be a 10 turn potentiometer fitted with a 10 turn dial. Alternatively, a long slider potentiometer, such as the Rivlin WS 150 (Ambit International) may be suitable.

Another tuning system involves the use of the IMI varicap control unit type 9932 (from Ambit International). This unit has 6 push buttons, each of which has its own potentiometer. Fine adjustment of the tuning can be made by rotating one of the push buttons when it has been pressed in.
AM tuning facilities may be achieved using a varicap system. In this case, the tuner will incorporate suitable AM tuning diodes in the AM section of the receiver.

## Image rejection

The main advantage of the more expensive front end units is the higher rejection of unwanted signals. This may be particularly important if one lives near to an ambulance, police or fire station which has an adjacent transmitter. The more expensive units (priced at around £12) can provide about 90 dB rejection of all spurious signals, whereas cheaper front ends (priced at about £6) may provide some 40 to 50 dB rejection.
Most VHF front ends employ a MOSFET in the signal frequency RF stage for low cross modulation whereas a bipolar transistor operates as a mixer; a few use a MOSFET as a mixer to minimise inter-


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modulation. This does not imply that a front end using a MOSFET is necessarily better than one employing, a conventional bipolar device since the one employing the bipolar type may contain several tuned circuits to increase the selectivity and therefore the rejection of spurious frequencies. The noise performance of a bipolar arrangement may be superior.

## Aerial impedance

The input impedance is usually either about 75 ohms for use with an unbalanced signal from a coaxial cable or about 300 ohms for use with a flat twin balanced feeder. If an aerial with the incorrect impedance is employed, the signal to noise ratio will be reduced somewhat. In an area of high field strength, this is not likely to matter.
A signal to noise ratio is quoted for most front ends, as a figure expressed in dB . In an area of low signal strength, noise performance may be important. However, minor differences are relatively inaudible. For example, a noise level of 6 dB is only just noticeably better than a 9 dB noise figure.

## Frequency coverage

Most units provide a frequency coverage of some 87 to 109 MHz , but all will cover the complete broadcast band with room to spare, especially at the upper end. If one selects a varicap front end, the
uppermost frequency of about 109 MHz will not be attained unless the tuning voltage can provide perhaps 20 V to 25 V . An upper limit of not less than about 101 MHz should be obtained with a 12 V and not less than about 103 MHz with a 15 V tuning voltage. A supply of 10 V or even less is adequate to cover the present broadcast band.

Various front ends require power supply voltages ranging from about 6 V to 12 V -no problems here. However, types requiring 12 V may not be so convenient for use in portable receivers.

## AGC/AFC

Most units do not have AGC facilities; they are generally unnecessary unless used very close to a transmitter. On the other hand, most have an AFC input and this can be most valuable if there is any tendency towards drift. Incorrect tuning results in increased distortion, especially at high modulation depth. Care must be taken to ensure that the AFC voltage from the receiver is suitable for the front end selected; some accept AFC voltages centred at 0 V , while others require an input which has a mean positive potential relative to the negative line.

## Commercial front ends

We will now comment fairly briefly on some front ends available from our advertisers. More formal specifications are given in Table 1.

|  | EF-5800HG | EF-5800HQ | EF-5600U EF-5602U EF-5603U | EC-3302U | 8319 | LP1186 | NT3302 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Tuning | varicap | varicap | varicap | varicap | varicap | varicap | ganged capacitor |
| Supply voltage | 12V | 12 V | 12 V | 9 V | 12 V | 8 V | 9 V |
| Supply current (mA) | 25-35 | 25-35 | 17 (max) | 16 (max) | 25 | $6 \cdot 1$ | 18 (max) |
| Gain (dB) | 44 | 37 | 30 | 22 | 32 | 30 | 25 |
| Noise figure (dB) | $8 \cdot 5$ | $7 \cdot 8$ | 7 | 7 | 5 | $6 \cdot 5$ | 8 |
| Tuning range ( MHz ) for tuning voltage | $\begin{aligned} & 88 \text { to } 108 \\ & 2 \cdot 5-16 \mathrm{~V} \end{aligned}$ |  | $\begin{aligned} & 87-109 \\ & 3-25 \mathrm{~V} \end{aligned}$ | $\begin{aligned} & 87-109 \\ & 2-20 \mathrm{~V} \end{aligned}$ | $\begin{gathered} 87 \cdot 5-108 \cdot 5 \\ 2 \cdot 3-18 \mathrm{~V} \end{gathered}$ | $\begin{gathered} 87 \cdot 4-108 \\ 2-17 \mathrm{~V} \end{gathered}$ | $87-109$ - |
| Input impedance (ohms) | 75 | 75 | 75 | $\begin{gathered} 300 \\ \text { (bal.) } \end{gathered}$ | $\begin{gathered} 75 \text { or } 300 \\ \text { (bal.) } \end{gathered}$ | 75 | $\begin{gathered} 300 \\ \text { (bal.) } \end{gathered}$ |
| Output impedance (ohms) | 300 | 300 | 75 | 300 | recomm'd load 150 | 75 | 300 |
| Image rejection (dB) | 78 | 97 | $\begin{gathered} 90 \\ (\min ) \end{gathered}$ | $\begin{aligned} & 45(\mathrm{~min}) \\ & 108 \mathrm{MHz} \end{aligned}$ | 56 | 40 | $45 \text { (min) }$ $108 \mathrm{MHz}$ |
| Thermal stability $25-55^{\circ} \mathrm{C}$ (kHz) | 180 | 180 | 100 | 150 | - | 300 | 150 |
| No of varicaps | $\begin{gathered} 6 \\ \text { back to back } \end{gathered}$ | $6$ | $\begin{gathered} 5 \\ \text { back to back } \end{gathered}$ | $\begin{gathered} 3 \\ \text { single } \end{gathered}$ | $\begin{gathered} 4 \\ \text { back to back } \end{gathered}$ | $\begin{gathered} 3 \\ \text { single } \end{gathered}$ | - |
| No of transistors | 4 | 4 | 4 | 3 | 3 | 3 | 3 |
| No of $10 \cdot 7 \mathrm{MHz}$ tuned circuits | 2 | 2 | 2 | 1 | 2 | 2 | 1 |
| $\begin{aligned} & \text { IF bandwidth } \\ & (\mathrm{kHz}) \end{aligned}$ | 300 | 300 | 300 | $\underset{(\mathrm{min})}{200}$ | 300 | $\begin{gathered} 200 \\ (\mathrm{~min}) \end{gathered}$ | 350 |
| Approximate size (body only) (mm) | $170 \times 20$ | $\times 24$ | $90 \times 61 \times 33$ | $66 \times 26 \times 15$ |  | $42 \times 31 \times 17$ | $74 \times 53 \times 49$ |

Table 1. Basic details of some front ends; full details should always be obtained from the data sheet before purchase

## NT3302UG

The Toko NT3302UG (Ambit International and from Doram Electronics) employs a ganged capacitor and is low cost. It incorporates a double gang capacitor for AM tuning with a range of 8.5 pF to 360 pF each. A similar Toko unit, the MT3302, has an AM tuning gang with 160 pF and 76 pF capacitance per section. A similar type is also available with a $3: 1$ reduction drive. These units incorporate a MOSFET input stage with bipolar transistors as mixer and oscillator. A single 10.7 MHz output tuned circuit is used. AFC facilities are provided.

## EC3302U

The Toko EC3302U (Ambit International) is an economical, compact varicap tuned front end. A MOSFET input stage is used with bipolar transistors as mixer and oscillator. As in the NT3302, the output is via the coupling winding of the $10 \cdot 7 \mathrm{MHz}$ output transformer. Although the EC3302U has less rejection of spurious signals than other Toko front ends, it provides an adequate performance in most locations for approximately half the price of the more expensive units. AFC facilities are provided.


## EF5800 series

The EF5800 front ends have been introduced fairly recently by Ambit International. They are unusual in that two RF stages are employed (both using MOSFET devices) with bipolar transistors as the mixer and oscillator. There are six circuits tuned by varicap diodes: an aerial circuit, two tuned circuits between the two RF stages, two tuned circuits between the second RF stage and the mixer and oscillator stage. A double tuned 10.7 MHz circuit is used in the output with a coupling winding providing the actual output.

The EF5800 units are available completely built and aligned, but are also available as a kit. While the kit is about $£ 2$ cheaper than the completely built and aligned unit, readers are strongly advised to ask themselves (more than once) whether they are equipped to construct this unit. If any doubt exists, it is well worth paying the extra. The EF5800 is available in two versions (see Table 1). The EF5800HG has a very high gain, but the EF5800HQ gives optimum rejection of spurious signals.

An EF5801 is also available with an FET buffered frequency counter output and is built on a double sided printed circuit board. The AGC range is also increased.

## EF5600 series

The Toko EF5600 series are high performance front ends with an extremely high rejection of spurious frequencies (minimum 90 dB ). This high rejection is obtained by the use of three signal frequency tuned circuits between the RF stage and the bipolar mixer. Five back-to-back varicap diodes (type BB-104) are used. An emitter follower output stage follows the $10 \cdot 7 \mathrm{MHz}$ double tuned circuits so the output impedance is low. A $0.01 \mu \mathrm{~F}$ blocking capacitor is used in the output circuit so that the output can be directly connected to any other circuit. The EF5600U has AGC and AFC facilities, whereas the EF5603U has neither. The AFC of the EF5600U is designed for use with ratio detector circuits giving a voltage centred at zero; if the AFC is not used, the AFC connection to the front end must be earthed. The
*A typical commercial unit with screening cover removed.
EF5602U has an AFC input suitable for use with the output from a quadrature detector, such as the CA3089E.

## 8319

The Larsholt 8318 (Ambit International) front end is manufactured in Denmark. It is a high performance unit with an especially good noise figure. There are two tuned circuits between the RF stage and the mixer. Both the RF stage and the mixer use MOSFET devices. Another unusual feature is the provision of separate inputs for 75 ohms (unbalanced) and 300 ohms (balanced). There is an AFC facility.

## LP1186

The Mullard LP1186 (Doram Electronics) is one of a series of modules. Although designed to be used with the LP1 $185 \quad 10 \cdot 7 \mathrm{MHz}$ IF module, the LP1186 can be used with any of the amplifier/demodulator integrated circuits. The AFC facility in this module is designed for use with ratio detector type equipment with an AFC input centred about OV. The gain is higher than that of the EC3302U, but the latter has better frequency stability.

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## FIII <br> <br> Demo <br> <br> Demo <br> 

Bill BOND G3XGP

MOST radio amateurs are aware of the advantages of narrow band frequency modulation (NBFM) transmitters in that little, if any, interference is caused to other services. Amplitude modulated transmissions, either AM or SSB, or CW can be picked up on a neighbour's television sets, a hearing aid or the Hi -Fi--the received RF overcones the bias on the input transistors causing demodulation of your signal. Increasingly, new amateurs coming on the two metre band modify or purchase equipment for frequency or phase modulation (FM or PM) gaining the advantage of an audio modulator running microwatts instead of at least half the power of the RF output stage required by AM. To those capable of receiving NBFM, ignition interference is largely eliminated as is the characteristic QSB or flutter found on / M stations.

## NBFM demodulation

Unfortunately not all amateurs receive NBFM under optimum conditions. With the average AM receiver, detection is done on the slope of the IF pass band (Fig. 1), the frequency of reception being offset from the centre of the carrier so that the signal moves in and out of the pass band of the IF and thus presents a variable voltage to the diode detector corresponding to the audio modulation. If the receiver has a fairly sharp edge to the pass band this method of reception can be very effective. However even the best of AM receivers fail where FM deviation is very small.
The solution is to employ a limiting amplifier with high gain to square the incoming wave. This eliminates audio modulation leaving only the frequency deviation which can be converted to an audio signal by a phase demodulator. In the past this has required a complex transformer to present a balanced input to a double diode detector such as the Foster Seeley or the ratio detector. Even a pulse counting demodulator gives poor recovered audio from NBFM. The phase locked loop is an excellent technique of recent development giving good AM rejection but expensive of space in isolated component form and expensive financially in integrated circuit form. Fortunately with the need to demodulate FM TV sound, cheaper, simple and very effective integrated circuits have appeared which combine a very high gain limiting amplifier with a double balanced demodulator. The latter is presented with a square wave, one side being at $90^{\circ}$ nominal phase difference resulting in demodulation of NBFM with an efficiency seldom achieved by older methods.
The author's NBFM demodulator design uses a high gain amplifier and a double balanced demodulator in a single IC with a minimum of external components to give an audio output of up to 0.8 volt, but pursues the matter a little further. Not


Fig. 1: The upper drawing shows a conventional $A M$ receiver tuned for reception of signals. The FM is converted to $A M$ using the receiver passband slope. The lower drawing shows a simplified quadrature detector. The relative phase between inputs of Tr1 and 2, a function of the input frequency when the combination of L1/C1 is near resonance, controls the current conduction angle through the common collector resistor R1. Recovered audio modulation appears across this resistor.
every amateur wishes to 'butcher' his equipment. The addition of accessory equipment should not alter the ability of the receiver to accept AM or SSB with equal facility. This discriminator includes an input stage which provides minimum loading on the output from the IF stages and, by including an audio stage, provides reception of NBFM through a separate loudspeaker. This enables both NBFM and AM stations to be received without adjustment of the receiver controls and also gives some indication of the amount of FM on an AM transmission. If the complete unit is made, only three connections are required: an input from the IF stage, and power supply take off. The unit has been designed for the two popular IFs of 455 kHz to 470 kHz , and $1 \cdot 6 \mathrm{MHz}$.

## Circuit description

A very high input impedance stage is provided by an FET (Fig. 2) capacitatively coupled to the last IF


Fig. 2: Circuit diagram. The prototype was constructed on a small piece of etched circuit board. The layout needs considerable care to avoid instability. Keep the input and output wiring of IC1 separated; use a screen if necessary. 11 will require a screening can.
stage, the input being best connected to the anode or collector through a capacitor of suitable voltage tolerance. Some readjustment of the last IF transformer slug will be required which should be minimum when the link between IF stage and discriminator is as short as possible.

Cl links IF to the gate of Tr which is held at earth potential by R1. R2 is the source resistance and provides a low impedance input via C2 to the IC, a plessey 432A. This has an input impedance of $1 \cdot 4 \mathrm{k} \Omega$ and incorporates an amplifier with a gain of over 50 dB followed by a balanced demodulator. Most of the capacitors surrounding the IC decouple the various stages. However, C 5 and Ll provide a $90^{\circ}$ phase shift between pin 3 and pin 5 at the IF, and both must be selected depending on the IF to be demodulated (see component list). The output from the balanced demodulator is fed via C10 to IC2.

## Construction

Because of the high gain of the preceding receiver IF strip, particular attention must be paid to abolition of ground loops-these can occur while connecting the demodulator unit to the receiver as well as in the component board itself. All conductors surrounding the 432A must be kept as short as possible; this particularly applies to the circuit decoupling capacitors.

The circuit can be constructed on either veroboard or etched PCB. The author used the latter method for the prototype. Whichever method is employed the input circuitry (Trl, etc) should be placed the opposite side to the output circuitry (L1, IC2, etc.).

Ll was wound on a standard 12 mm pot core as used in many IF transformers. For 455 kHz wind 80 turns of 40 gauge enamel covered wire. C5 should be about 1000 pF . For $1 \cdot 6 \mathrm{MHz}$, wind 50 turns of similar wire with C 5270 pF . It is essential to screen Ll in a metal can to avoid instability. For those without 12 mm pot cores readily to hand, it might be worth experimenting with converted IF transformers. Ll and C5 should be series resonant at about the the receiver intermediate frequency. In operation, L1 is 'mistuned' so that the phase shift across C5, normally $180^{\circ}$ at resonance, is reduced to $90^{\circ}$. With standard high $Q$ IF transformers, this will occur within a few kHz of nominal resonance; the recovered audio
depends on the $Q$ of the LC combination. The higher the Q , the higher the output for a given deviation. Adequate results may be obtained by breaking open an old IF can, removing one winding and tuning capacitor, and using the remaining winding for L1 and the associated capacitor for C5. Slight adjustment may be necessary to compensate for the capacitative loading of ICl in the 'new' application.

## components list

```
Capacitors
    C1 1000pf
    C2 2000\muF
    C3, 4, 7, 8, 9 0.1 \mu\textrm{F}\mathrm{ ceramic}
    C5 \begin{array}{c}{1000\textrm{pf}\mathrm{ for }455\textrm{kHz}}\\{270\textrm{pf for 1.6 MHz}}\end{array}}\mathrm{ See text}
    C6 33\mu\textrm{F}
    C10 2.2\muF
    C11 2200\muF
    C13 4.7,tF tantalum
    C12 100\muF
```

Resistors
R1 $470 \mathrm{k} \Omega$
R2 1 kS :
R3 $1 \mathrm{k} \Omega$
VR1 100kS2 linear

## Coils

L1 Standard 12 mm pot core as used in many IF transformers. Wind 80 turns 40 gauge enamel covered or Litz wire for 455 kHz , or 50 turns similar wire for 1.6 MHz . Screening can essential. See text.

## Semiconductors

IC1 432A Plessey
IC2 LM380

## Setting up

Connect a 12 V supply to the unit; the current drawn should total about 30 mA . Feed an FM IF signal to the unit at the mean operating frequency and adjust L 1 for maximum output. A fairly accurate adjustment may be made by disconnecting the aerial from the receiver and tuning L1 for maximum white noise in the loudspeaker. However, really accurate adjustment can only be made with a NBFM transmission.

MAINS POWERED
VARIABLE LENGTH SUSTAIN RHYTHMS MAY BE MIXED

2 PRINCIPAL TONE COLOURS 2 OCTAVES OF ACCOMPANIMENT HIGHLY COMPACT: DIMENSIONS 20" $\times 9^{\prime \prime} \times 2 \frac{1}{2}$ 3 OCTAVES FOR MELODY LINE: 3 TONE CONTROLS RHYTHM SECTION: BASS, SNARE DRUM \& CYMBALS

PHONO OUTPUT PERCUSSION EFFECTS AUTOMATIC CHORDING VARIABLE DEPTH VIBRATO VARIABLE RHYTHM SPEEDS EASY SINGLE BOARD CONSTRUCTION INTERNAL 1 W AMPLIFIER TO DRIVE LOUDSPEAKER DESIGNED TO BE PLAYED WITH 1 FINGER OF EACH HAND

[^2]
## FEATURED IN JUNE 177

.EL $\left\{\begin{array}{l}\text { FDDTBA } \\ \text { SDDASL } \\ \text { DEDDIA }\end{array}\right.$
ON-SCREEN SCORING \& SOUND EFFECTS

THE General Instrument AY-3-8500 IC has two positions on the games switch that provide a square target for use in shooting games. In the first position the target bounces round the screen in an invisible box. In the second position the target crosses the screen from left to right. When the reset button is pressed the screen is blanked except for the target. After 15 shots the score appears to indicate the number of hits out of 15 .

SHOOT is an add-on unit which allows an aiming mark to be placed on the screen and moved around by joystick control to chase the target. If the firing button is pressed when the aiming mark is on target a point is scored. A variable timer allows the aiming mark to be placed on the screen for preset times.

## Circuit description

The circuit of the unit is shown in Fig. 1. The unit obtains its supplies from the main 8500 PCB. The sync from the 8500 is fed to pin 5 of IClb and to Trl. IClb fires on each line sync and controls the horizontal position of the aiming mark, its range being approximately 15 to $55 \mu \mathrm{~s}$. The components C5, R5, D2 determine the width of the mark, the values on the circuit giving a width of $0.5 \mu \mathrm{~s}$. Trl is used to separate the frame sync from the line sync and provides a trigger pulse each frame to pin 12 of ICla. ICla controls the vertical portion of the aiming mark and is variable over the range approximately 5 to $18 \mu \mathrm{~s}$. Components C3, R4, D1 determine the height of the mark; the values shown give a height of 3 lines. S1 is the fire switch, debounced by IC3a and b. IC3 output from pin 10 is fed to the shot input of the AY-3-8500 incrementing the shot counter each time the trigger is pressed. C6, R8, D3 determine the width of the fire window. C6 can be increased to make the window wider which makes the game easier.

## Gating

ICla and b gate IC4a together with the output from IC2a. When S2 is pressed IC2a fires for a period determined by VR3 opening gate IC4a. The output of IC4a is inverted by IC4b, used to produce the spot on the screen, and also gates IC4c. The ball

## D.S.COUTTS

output from the AY-3-8500 and the fire window from C6 are also fed to IC4c. When a hit is detected by this gate, IC2b fires and outputs a pulse to the 8500 hit input. This increments the hit counter, outputs a tone and blanks the target for the duration of the hit pulse.

There are therefore seven connections between the 8500 main PCB and SHOOT: $0 \mathrm{~V},+8 \mathrm{~V}, \mathrm{SYNC}, \mathrm{BALL}$ OUTPUT from the main PCB and shot, hit and AIMING MARK from SHOOT.



## Construction

Actual mechanical construction will depend on the size of the joystick used; a typical layout is shown in Fig. 5. As standard joysticks used for radio control may have $5 \mathrm{k} \Omega$ pots fitted, ensure that the type obtained can be modified to take a $25 \mathrm{k} \Omega$ and a $50 \mathrm{k} \Omega$ pot. If one is obtained with a tubular operating arm, S1 may be mounted in a piece of dowel and fitted on top of the operating arm.

## components list



Fig. 2 (left) : PCB copper side shown acfual size. The board is secured to the base of the plastic housing by two 6BA bolts inserted through the drillings indicated on the board. Circuit dlagram (Fig. 1) is shown on page 283.

Fig. 3 (right) : PCB component layout. DIL sockets should be used for all ICs unless the usual CMOS soldering procedures are followed. Devices should be left in the conductlve foll or foam until ready for insertion; plns 2 anc 14 of the MC14528 are especially susceptible to statlic damage since these do not have the normal hoput protection diodes.


When the mechanical parts have been mounted in a suitable box, begin assembly of the PCB. Fit the wire links as shown by the single lines on Fig. 3 and fit the other components. As CMOS devices are used in the system it is advisable to fit IC sockets and leave insertion of the ICs until construction is finished. If you do not use sockets then use an iron with an earthed tip for soldering. When the PCB is complete fit it into the box. Fig. 3 also shows the connections to the board.

## Connection

A seven-pin plug should be fitted to the cable to join the unit to the TV games unit with a matching socket in the latter's box. The connection points on the 8500 PCB are shown in Fig. 4. The sync output is taken from pin 16 on the 8500 IC and the marker spot lead connects to the end of R9. The +8 V output

Fig. 4: Interconnection diagram for use with PCB featurzd in the June 1977 Telegames unit. Although the prototype SHOOT module was connected to the main unit by flying leads, readers are lecommended to fit 7 -pin DIN plugs or other multiway connectors to both master and slave games unit.


Fig. 5: Photo of the prototype with cover removed and shown the wrong way round. The pan pot assembly fits towards the opposite end from the PCB and not as indicated in the picturel The pan pot assembly was originally intended for use with radio control apparatus; the actual potentiometers are standard components. The joystick assembly should be obtainable from radio control specialists.


Fig. 1: Circuit diagram. The text referring to this occurs on page 280. IC3c and d are unused gales, the inputs of which must be grounded or tied to the VCC line. They must never be left floating. The power supply to the SHOOT module PCB is taken from the parent board in the Telegames unit. Note: VR4 should be shown as $22 k \Omega$ as in components list.
comes from the regulated supply. The other connections are on pins at the edge of the PCB.

## Setting up

After checking the unit carefully for shorts, etc, connect the SHOOT adaptor to the main unit. Set VR3 to maximum resistance. Select rifle game R1 and auto mode when the target should appear on the screen. Push S2 and adjust the position of VR1, VR2 and the presets VR4, VR5 to get the aiming mark on the screen. Remember that S 2 has to be
pressed occasionally to keep the mark visible. When the mark has appeared on the screen adjust the pots and presets until the mark can be swept over most of the screen by means of the joystick. When this is satisfactory try out the fire button when a tone should be obtained each time the target is hit (N.B.-when moving the stick with the target moving fast it is easy to think you have a hit which hasn't registered when in fact a slight delay in pressing the fire button has caused you to miss). It is advisable to use slow speed for testing; further, the fire window can be widened temporarily by connecting another capacitor across C6.

## Mind your language

Right now it's holiday time-going abroad or maybe you are abroad? Bet you wished you could talk the local lingo and that you had paid more attention in those language classes when you were at school?

A method of language tuition that has been popular for some years now is the language laboratory which of course is hardly portable. But now Philips have introduced a new language trainer the AAC4000 based on their audio active comparative (AAC) system and it really can be carried on the shoulder. In laymen's terms, AAC means that the user listens to sentences of tuition prerecorded on the lower half of the training compact cassette. There are long pauses between the sentences of tuition and during the pauses the student repeats the sentences, which are then recorded on the cassette's upper track. Subsequently the student can replay both tracks at the same time so that he can compare his own pronunciation with the teacher's. If he wishes to try again the student can erase his previous performance but the tuition track is automatically protected against erasure.

The separate tracks can also be used for other applications such as slide and sound synchronisation.

Philips Electrical Ltd., Century House, Shaftesbury Ave., London WC2. Tel: 07-437 7777 .

## It's in the bag

It's not every reader who can make his own printed circuit board and make a good job of it-the usual trouble is the mess created by the etching solution, it seems to get just about everywhere except onto the board where it's. needed. If the board and solution were contained in a special bag the major problem would

be overcome-no more mess. Verospeed have adopted this idea with their new Etching Kit which comprises all that is needed to remove copper in a clean, safe and simple way from ten Eurocard-size boards.


A special neutraliser is also included which ensures environmentally safe disposal once the kit is exhausted. An etch resist pen and transfers, together with a copper cleaning block completes the kit, which is guaranteed, says Verospeed, to be available by return post. Each kit is priced at $£ 4.30$ plus VAT and has code number of 22-4410G.

Are you the type that likes a bit of noise, especially when your car is about to be pinched or your house is being burgled? You are? Then how about a miniature audible warning device that takes very little current and can be satisfactorily operated from 5 to 30 V DC or 20 V AC. A

choice of two continuous signals, high or low frequency is available from the one unit, which is priced at $£ 3.52$ plus VAT. Verospeed order number is $41-4642 \mathrm{~J}$.
Verospeed Service, 10 Barton Park Industrial Estate, Eastleigh, Hants. SO5 5RR. Tel: 0703618525.

## Making waves

Time was, when, if you required a specifically shaped signal, the equipment you required tended to be a bit cumbersome. With the introduction of integrated circuits this exercise has now become a little mundane, and IC's can be obtained which give a sine, square and triangular waveform output over a fairly large frequency spectrum. Lascar Electronics now market a waveform generator module comprising IC, peripheral components all mounted on one board, and whose frequency on all three waveforms extends from 20 Hz to 20 kHz . Operating on a supply voltage of $\pm 5$
to $\pm 15 \mathrm{~V}$, this CB-FG8038 generator has typically $1 \%$ distortion, duty cycle variations of $2 \%$ to $98 \%$ and power dissipation of 750 mW .
Output frequency is set by an ex-

ternal 10 k ohm pot, and by adjustment of the on-board duty cycle pre-set pot, sawtooth, ramp and pulse waveforms can also be obtained.

The Cirbloc module is assembled onto a PCB measuring $76 \times 48 \mathrm{~mm}$ and is priced at $£ 7 \cdot 16+$ VAT.
Lascar Electronics Ltd., PO Box 12, Module House, Billericay, Essex. Tel: 027743394.

## Take me to your leader

No it's not a copy of Penthouse or the latest fashion magazine for bathers at St Tropez - it's - wait for it, the latest catalogue from Eagle International. I can just read your minds now, what could that strange cover possibly have to do with Electronic components and HiFi gear? Well the answer is nothing, but never-the-less what it has got to do with is extremely interesting. The

picture, which incidentally is produced by NASA is supposed to convey to anf alien being all that he
needs to know about Homo Sapiens (Man). What do you mean you don't see it? Its quite simple really, the 'rays' represent our sun in respect to the closest radio-emitting stars (pulsars). Also shown are the relative positions of the planets in our system and the size of the vehicle (spacecraft) compared with its creators. Much more information is encoded using the binary system, all of which could be deciphered by any intelligent lifeform. So, if there are any readers who classify themselves as 'Intelligent Lifeform' please drop PW a line and explain it all to us!
Incidentally the Eagle Catalogue is pretty good too.
Eagle International, Precision Centre, Heather Park Drive, Wembley. Tel: 01-902 8832.

## $1 \frac{1}{2}$ billion ohms

A new range of BiMOS Op-Amps incorporating MOS/FET inputs, COS/ MOS outputs, and having low drift figures of $6 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ have now been introduced by Mogul Electronics.

Notated the CA3160, output current is in the order of 20 mA , bandwidth is

4 MHz , input impedance is typically high, being in the order of $1.5 \times 10^{12}$ ohms and the output voltage is capable of swinging to within 10 mV of either supply rail.

The CA3160 family is said to be ideal for interfacing with COS/MOS, and all devices operate within the temperature range of $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$. Packages obtainable com-

prise a standard 8-pin TO99 package or a DIL can device with TO99 lead configuration preformed to 8-pin DIL layout.
This device is manufactured under the YES Electronics label which is part of Mogul Electronics Ltd., 273 High Street, Epping, Essex. Tel : 037877366.

## Thidytray

You know it always happens, you put down the screwdriver and five minutes later its gone-walked off. You swear and curse and accuse everyone around you and then look on the bench, and then, yes you've guessed it, the screwdriver has mysteriously returned. Well these things are sent to try us, but wouldn't it be nice if we could keep all our tools together, know where they are, and
know that they won't walk off, or roll off or just plain dissappear.

Obviously a firm by the name of Cannon have had the same sort of problem that you and I suffer from, because they have recently introduced a rubber mat that's flexible, non-slip can be draped over an irregular surface or laid flat on a polished table top. It has internal dividers, and a 1 inch lip around the tray provides for a variety of small tools.

Called the Autotidy, the price is £1-38 and can be obtained from most DIY stores.


## ON RECENT DEVELOPMENTS

## Citrus power

How about shoving an orange into your transistor radio next time the battery runs flat? Not quite a practical proposition, but lemons were used as a source of power at a recent electronics exhibition in London. This power was then stepped up to give a spectacular display.

The "secret" lies in the fact that if you unkindly stick two dissimilar metals into the acid contents of a citrus fruit, such as an orange or a lemon, a voltage and current are generated. The output is, of course, quite small, but your ageing scribe has had a CMOS oscillator running from a single lemon-quite an impressive demonstration for when friends call.

Perhaps, with power conservation presently in everyones' minds, farmers will be encouraged to grow special "electrical" varieties of fruit. Imagine scouring neighbourhood greengrocers in search of a 3 volt grapefruit!

## You can't beat it

A chap called Khayyam wrote, "The moving writes and having writ . . ." etc. In modern times, however, the moving finger moves and rests lightly on top of a digital wrist watch. That's so that the owner can check his pulse rate-of course!

Just out in the States is a digital watch with a tiny two-sensor button on the top, just below the readout face. One little sensor is a light transmitter while the other is a light receiver. Resting the finger directly across the two, switches on the pulse counting sections of the watch. The tiny transmitter emits infrared light. This travels into the finger and is reflected back again to the receiver sensor. Blood in the capillaries of the finger expand and contract very slightly as blood is pumped through them. It is this tiny change which alters the amount of infrared light received by the receiving sensor. The result is a series of tiny infrared pulses triggered or caused by the actual body pulse itself.

Individual pulses are amplified and passed to a control counter then to a count-digit converter and so to the
digital readout which shows the pulse rate directly. Since the digital watch, has a very accurate seconds pulse (from the watch oscillator) the tiny microcomputer in the watch can accurately measure the pulse and give a direct readout in beats per minute. Sad to say you cannot buy one yet and the price in the US is around 500 dollars for a stainless steel case version. The watch may be accessed or interogated 25 times every day for a whole year before the batteries need changing. Presumably this watch is aimed at impulse buyers.

## Light time piece

Talking electronic watches, I note that a German company is to market two quartz-driven clocks with accuracies of $\pm 30$ seconds a year. The difference with these timepieces is that they have built in solar cells which keep the batteries charged. One is designed to function in daylight (not necessarily sunlight) which is interesting. Perhaps transistor radios will follow suit when these cells become cheaper and more efficient.

And talking about solar cells reminds me that President Carter's new energy programme has some support already. For example a knitwear factory is currently under construction in the US which will eventually employ some 300 people. Its power requirements are $1,400 \mathrm{~kW}$ and this is to be provided entirely by the sun. At last solar power is really getting down to the knitty gritty.

Carrying on from a previous Ginsberg when I mentioned a professor who had an all-electric car, i see that a great deal of work is being done in this field. Two cars are under investigation by one company. One car has electronic controls while the other utilises a huge flywheel for better acceleration. These are both four-passenger models with a top speed of 55 mph and a range, at present, of 75 miles. Some four million dollars is invested here in contracts which run for two years.

## Yall marked display

Ever heard of LETI? No, not a little Yhetti but the laboratory for electronic
and computer technology attached to the French atomic agency (yeah, I think it's easier to say LETI, too!). Apparently a group of researchers there have discovered that by taking an organic electrolyte, plus a drop of silver, they can make readout cells which, they claim, are better and cheaper than liquid crystal types. In their normal or "off" state, these cells are transparent. They change to opaque in some 50 ms when a 1 V pulse is applied across them. The application of this pulse causes a very thin layer (typically 50 Angstroms) of silver to "plate" the electrodes. This plating will stay in position for around 10 minutes before the cell begins to go transparent again so it might be argued that these devices have a ten minute memory. A pulse can be applied which will clear the cells immediately.

## Cap that!

Discrete capacitors are still used by the millions in almost every conceivable electronics application. One company in the capacitor field has come up with a good idea. They manufacture foil and film capacitors and they've brought out some devices called Multipacitors. In a single unit package, this company houses a number of different capacitors offering a $30 \%$ saving in space compared with using separate discrete components. The trick is to take tapping points of the foil winding to offer different capacitance values within the same package. Instead of buying four components, you buy only one! One package offers, $0.47 \mu \mathrm{~F}$ at 400 V wkg plus $1 \mu \mathrm{~F} 400 \mathrm{~V}$ wkg plus $2 \mu \mathrm{~F}$ at 200 V plus $4 \mu \mathrm{~F}$ at 200 V wkg.

I note with interest that the latest digital display devices have drivers and decoders built in. All you need do is feed the beast direct from a four line BCD output. They're TTL compatible, too. Whatever will they integrate next?



# N.C.HELSBY,MA, MIEE 

THE advent of quartz crystal oscillators in watches, and the regulation of conventional clocks and watches, poses a thorny questionhow does one check their performance? Clocks driven by the mains supply, either digital or mechanical, indicate to a high accuracy only if viewed over a period of months or years, compared with a quartz watch, since they may be a number of seconds wrong at any particular moment-and by this time there may have been a power cut! The Post Office Speaking Clock could prove expensive to an enthusiast-and if the telephone is not next to your favourite grandfather clock how do you set the exact second? Greenwich time signals from the BBC can turn out to be very elusive just when you need them.
Standard time and frequency signals have been transmitted for many years on both short and very long waves. The most convenient of these for time checking purposes in this country is the 60 kHz transmission of Rugby MSF. This station transmits seconds and minute markers as well as a call-sign which enables identification of the hour. Recently, at the end of each minute, a time code was introduced in which the hours and minutes are signalled in a binary code; it is possible to construct a digital clock which sets itself by means of this code. However for many purposes the seconds and minutes markers, which are easily recognized by ear, are all that is required for setting and checking clocks or watches.

## Atomic transmission

The time by which we live must clearly be related to the rotation of the Earth on its axis but, even when all possible corrections have been made, the rate of this rotation fluctuates unpredictably. Also the Earth is known to be slowing its angular rate due to tidal friction (admittedly amounting to only a millisecond a century), thus in principle to arrive at a standard invariable unit of time such as the second another definition is required. In 1964, at a meeting in Paris, the International Committee of Weights and Measures adopted as its tentative standard an atomic transition, specifically a hyperfine transition in the atom of caesium ${ }^{133}$. To keep the time of day measured by atomic clocks using this definition of the second (the time for $9,192,631,770$ transitions) in step with the time scale referenced to the rotation of the Earth about its axis corrections by means of the addition or subtraction of a complete second are occasionally carried out, usually on either June 30 th or December 31st. Clearly whilst checking the performance of accurate clocks it is useful to know
when these corrections have been made, as they will be applied to Greenwich time signals as well as the other accurate sources of time.

## Rugby MSF

The Rugby transmission gives an indication of when such leap seconds (or the reverse) have been included. The normal seconds markers are formed by a complete $0 \cdot 1 \mathrm{~s}$ break in the 60 kHz carrier. The minute marker is easily identified as it is a 0.5 s break, apart from the first $0 \cdot 2 \mathrm{~s}$ or so which contains the time code. In each case the start of the break in the carrier represents the exact time, so that a receiver which "bleeps" when the break occurs sounds sinilar to Greenwich time signals as transmitted by the BBC, apart from a slight croak at the minute due to the presence of the time code: except for servicing periods (usually a few hours on the first Tuesday in the month) the signals are available 24 hours a day.

The difference between the time scale related to the rotation of the Earth and the atomic scale is given in units of 0.1 s by means of an extra pulse (where a pulse is no carrier) following each second from second number 1 to, at most, second number 7 , for the atomic scale lagging behind the rotation of the Earth. If the atomic scale is ahead similar pulses following seconds 9 to up to 15 are inserted.

Thus a sudden change from, say, an extra pulse after each second from 9 to 15 inclusive to an extra

 transmitted by Rugby MSF. The BCD time code occurs at the start of each minute pulse train. Relationships between. atomic time and earth rotational time notated by appearance of dauble pulses in either seconds 1 to 6 (atomic time behind earth) or in seconds 9 to 15 (atomic * time ahead of earth).

Atomic scale ahead of rotation of earth by $0-2$ seconds
pulse after each of seconds 1,2 , and 3 would show that an extra second had been inserted into the atomic time scale, which has therefore jumped by this amount relative to all other clocks and watches; these, if their accuracy merits it, must be readjusted to follow suit. A change from, say, an extra pulse after each of seconds 1,2 and 3 to one after 1 and 2 merely indicates how the rotation of the Earth is going relative to the atomic scale. In other words, the error does not concern us with regard to checking other clocks until the whole second correction is made, which is apparent from the sudden switch in positioning of the double pulses. See Fig. 1.

## The receiver

The usual method of listening to a telegraphy signal, for example morse code transmitted by the keying on and off of a carrier, is to heterodyne or beat a local oscillator with the signal to produce an audible tone. This is one of the functions of the BFO (beat frequency oscillator) found on short wave receivers. However to make the gaps in a signal audible an audio oscillator is required to be switched off by the signal and turned on in its absence. This is slightly unusual, and, incidently renders, in the case of the Rugby signal, the morse code "MSF" sent


Fig. 2: Circuit of the atomic time receiver.
twice just before each hour unreadable, but gives the short bleeps most. convenient for setting clocks.

Starting from this end of the receiver (Fig. 2) any audio oscillator capable of being switched on and off will suffice--the one shown uses the familiar 555 timer, the output of which may connect to a small loudspeaker or earpiece such as the standard telephone type available from surplus suppliers. Note that a resistor is shown in series with the timer output to limit the current in the loudspeaker or earphone. It was found that 470 ohms was about the lowest value that could be used without endangering the normal operation of the 555 when feeding such loads. IC5 is turned on and off by the output of the receiver. therefore when no carrier is present the reset line (pin 4) is taken high by the output of IC4 and the output of the timer causes a "bleep" from the earpiece, Fig. 1.

The multiplier is used as a squaring circuit. The signal is multiplied by itself to produce a DC output, as well as a double frequency output which is removed by the following amplifier. The signal is fed into the two balanced inputs of the multiplier via the secondaries of a transformer T1. The bias point for each pair of inputs of the multiplier and the setting of the multiplier "tail" currents are obtained from a chain of resistors including R6, R7 and R8. The primary of T1 is tuned by means of C6 to 60 kHz . This transformer is wound on a Mullard type RM6 core more details of which will be given later. Trl is a common emitter amplifier having Tl as its collector load and its input fed from the secondary of the ferrite rod aerial. The primary of the aerial is tuned by Cl.

## Schmitt characteristic

Thus when a 60 kHz signal is picked up and amplified by Tr 1 it is applied to both inputs of the multiplier. This causes pin 6 of ICl to go slightly positive and pin 9 slightly negative each with a 120 kHz signal superimposed. IC2 then provides a great deal of amplification-about 625 times the differential input signal at DC. C11 and the amplifier's own frequency response ensure that the 120 kHz component is removed. The stability of resistors R1l and R12 is important as any change in the value of one of these alone will cause a change in the output of IC2. Providing R11 and R12 change by the same amount there is no output change so they should preferably have matched temperature coefficients. The receiver will work with carbon film types here but if excessive drifting with wide temperature changes is experienced metal film type (not necessarily metal oxide types unless selected for matched coefficients) eliminate the drift.

## DC level

VRI is used to set the DC level at TPl with no input signal at the aerial (i.e. aerial shorted). A 60 kHz signal causes the level at TP1 to fall. while the following amplifier, IC4 has positive feedback to give it a Schmitt characteristic. A reference level for IC4's negative input is obtained from the bias chain that feeds the multiplier. The purpose of this stage is to make sure a definite change in signal level is required before an output swing from high to low or low to high can occur, hence obtaining the best possible performance in the presence of noise.
components list

| Resistors |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| R1 | 22kS |  | R13 | $1 \mathrm{M} \Omega$ |
| R2 | 33kS2 |  | R14 | $1 \mathrm{M} \Omega$ |
| R3 | 180S2 |  | R15 | $1 \mathrm{M} \Omega$ |
| R4 | 1-8k $\Omega$ |  | R16 | 15k 2 |
| R5 | 2.7kS |  | R17 | 180kS |
| R6 | 1.5kS |  | R18 | 2.7kS |
| R7 | 1-8kS |  | R19 | 1kS2 |
| R8 | $2 \cdot 2 \mathrm{k}$ / |  | R20 | 47082 |
| R9 | 10052 |  | R21 | 150,2 |
| R10 | 100S2 |  | R22 | 10kS |
| R11 | $1 \cdot 5 \mathrm{k} \Omega \pm 2 \%$ metal film |  | R23 | 82kS |
| R12 | $1 \cdot 5 \mathrm{ks}$ ! $2 \%$ metal film |  |  |  |
| VR1 200s2 cermet trimmer | 200s2 cermet trimmer |  |  |  |
| All resistors $\frac{1}{6} \mathrm{~W} \pm 5 \%$ carbon film (except R11 \& 12) |  |  |  |  |
| Capacitors |  |  |  |  |
| C1 | 1.5nF Polystyrene | C8 | 56pF | Polystyrene |
| C2 | 100nF Polycarb. | C9 | 100 nF | Polycarb. |
| C3 | $32 \mu \mathrm{~F}$ or $47 \mu \mathrm{~F} 10 \mathrm{~V}$ | C10 | 100 nF | Polycarb. |
| C4 | 100 F F 12 V | C11 | 1 nF P | olystyrene |
| C5 | $32 \mu \mathrm{~F}$ or $47 \mu \mathrm{~F} 10 \mathrm{~V}$ | C12 | $10 \mu \mathrm{~F} 6$ | $3 \vee$ Tantalum |
| C6 | 4.7nF Polystyrene | C13 | 10 nF | Polycarb. |
| C7 | 100nF Polycarb. | C14 | 10 nF | olycarb. |

Semiconductors

| Tr1 | MPS-HOS or BC183 or ZTX 300 | IC3 | 741 |
| :--- | :--- | :--- | :--- |
| ZD1 | BZY88C5V1 $5 \cdot 1 V$ zener | IC4 | 741 |
| iC1 | N5596 or MC1496G or SG1496T | IC5 | NE555 |
| IC2 741. |  |  |  |

Miscellaneous Signal strength meter $100 / 1 \mathrm{~A}$ FSD. Ferrite rod aerial LW only, type LW5FR (Denco). T1, RM6 R Inductor core plus parts (Mullard)

## AGC action

Finally AGC is obtained from IC3. The output of IC3 controls the current in Trl, and hence the gain of Tr 1 , by determining the voltage at the bottom end of R4. The positive input terminal of IC3 is taken to a reference obtained from a convenient level on the bias chain. IC2 attempts to adjust the gain of $\operatorname{Tr} 1$ in order to obtain the same level at TP1 as the reference level. Thus if the signal level falls, the voltage at TP1 rises causing current to flow into R14. The output of IC3 then falls slowly increasing the gain of Trl, which returns the voltage at TPl to the reference level (about $2 \cdot 9 \mathrm{~V}$ ). When the carrier breaks for short periods, as for the seconds or minutes markers, IC3 starts to increase the receiver gain but is prevented from doing so very much by C12.

Note that on switch-on the receiver is in the minimum gain state with the output of IC3 high and gradually falling until the desired signal level is obtained. This normally takes only a few seconds, during which time the tone generator is continuously "on". The voltage to which the output of IC3 goes is a useful measure of signal strength and if desired a small meter measuring this level may be fitted.

## Construction details

The PCB is shown in Fig. 3. Note that there are a total of four links on the circuit board to be fitted in addition to the components. Additionally, points 16 and 17 should be linked for a continuous output, if a switch is not fitted, as may be required if the receiver output is used to control circuits other than the on-board 555 timer.


Fig. 3 (above): PCB copper side shown actual size.
Fig. 4 (below): PCB reversed showing component locations.


TMB851

Note that the trimmer potentiometer VR1 is a minature multiturn type. It should have a cermet track for good resolution-the wirewound types of low resistance are difficult to adjust finely due to the incremental resistance of each turn of the track as the wiper traverses it. The ferrite rod aerial specified is a standard transistor radio type available from Denco (Clacton) Ltd. When ordering, specify a long wave coil only as this reduces the cost. The resonant frequency of the aerial can be trimmed by sliding the coil along the rod from its nominal position. It is important that the coil should not be near the end of the rod in its final position as the rate of change of inductance with position is too fast for good stability.

The coil on its former as received may be a loose fit on the rod, in which case it is a simple matter to first wind onto the rod sufficient "Sellotape" to achieve a firm but readily adjustable fit.

The transformer Tl has been designed using the Mullard RM6 core. The RM series of cores use bobbins to which pins are attached, so that a coil may normally be terminated on these before the two halves of the core are fitted. This is done for the primary only in our case, and the secondaries are left with flying leads to be connected directly to the PCB. The following instructions should be adhered to carefully, remembering that a l turn error is significant when the total number is small.
l. Strip enamel insulation from the wire end and attach the wire to the pin marked "start primary". Do not exert undue pressure on the pins which are mounted on a fragile base.
2. Fix the bobbin to a suitable mandril such as a twist drill; just over 6 mm diameter is required whicii may be obtained by winding a thin layer of tape on a 6 mm drill.

Wind on wire evenly, top of bobbin moving towards you, with the pins facing to the right.
3. Complete 59 turns, cut wire, strip insulation and attach to "finish primary". Cover winding with 6 mm wide tape.
4. Starting with wire in "slot 1 " leave 30 mm flying lead, identified as "start" by fitting sleeving or other means, and wind on 7 turns exactly turning in same direction as before. Form a loop 30 mm long and twist the loop together: leave this a flying lead also in "slot l". Continue to wind on a further 7 turns exactly, take the remaining end across to "slot 1 " and leave 30 mm flying lead. Wind on a turn of 6 mm tape to fix the winding to the bobbin.
5. Repeat the procedure 4 above using "slot 2 ", which is on the same flange of the bobbin, facing "slot 1"

The bobbin now has six flying leads which are in two groups of three, each of which has a start, centre tap and finish. Wire these to the PCB as indicated on the circuit.
After winding the coil, check that the faces of the two halves of the core are free from dirt and carefully assemble them round the coil. After fitting the clips make sure that the core halves are aligned at their edges. Fit the adjuster and screw it down 5 turns to the nominal position, using a screwdriver which fits the small slot-if necessary modify or make a screwdriver to fit (eg from a knitting needle) rather than risk damaging the adjuster.

Note that if the phasing of one of the secondary windings is reversed the output at TP1 will increase

with increasing signal causing the AGC to latch the circuit up. If a mistake has been made in winding T1, all is not lost-simply reverse one of the secondary windings (eg swop the wire from 10 with the wire from 12 ).

## Setting up

The bandwidth of this receiver is about 300 Hz which is of course much narrower than that of normal broadcast types. This is just sufficient to allow correct reproduction of the time code so that all the information transmitted is available at the receiver output and a wider bandwidth would merely admit more noise. Hence if a signal generator is used it will probably need to be set to 60 kHz using a frequency counter: a 1 per cent error in frequency is 600 Hz and therefore significantly off-tune. Fortunately the receiver can easily be adjusted to the "off-air" signal. First short-circuit points 2 and 3 to which the aerial secondary is connected. Switch on the receiver.

Depending on the initial setting of VR1 there will be either no output, or a continuous tone. If there is no output, turn VR1 clockwise until the continuous tone just appears. If there is already a continuous output on switching on turn VR1 anti-clockwise until the tone disappears and then clockwise about ${ }_{4}$ turn so that the tone has just been switched on again. There is of course no variation in the sound of the tone output-it is simply on or off. Un-short the aerial and make sure that the coil on it is in its nominal position. The signal will probably now appear.

The ferrite rod aerial is directional and maximum signal is obtained when the direction from which the received signal arrives is at right angles to the line of the aerial rod. If the rod is turned away from this direction a point may be found where the signal becomes noisy (the timer output intermittent instead of a pure tone). Reduce the noise to a minimum by tuning $T 1$ and adjusting the aerial coil. If a voltmeter is available, connect it between 0 V and the output of the AGC amplifier IC3. Adjust T1 and the aerial coil for a maximum reading on the voltmeter. Note that adjustments must be made slowly, due to the slow response of the AGC-otherwise the person making the adjustment becomes part of an unstable feedback loop!


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## s-Decnolegy



THERE are some circuits which can be used for a wide variety of applications, of ten with only minor modifications. This month's S-Dec project falls into that category. I can think of four applications (see later), but readers will doubtless think of others.

## Circuil descriplion

The project comprises a free-running multivibrator made up from transistors Trl and Tr2. The output from this oscillator is then fed directly to the base of $\operatorname{Tr} 3$, which in turn feeds the amplified signal to the loudspeaker in its emitter lead.

The frequency of a free-running or astable multivibrator is dependent upon the values of resistance and capacitance employed. In the circuit (Fig. 1) these components are the two capacitors, Cl and C 2 , and the resistors between the bases of the transistors


Fig. 1 : Frequency-to-light converter circuit diagram. LDR1 controls C1 charging rate and hence the frequency of the oscillator.
$\operatorname{Tr} 1 / \operatorname{Tr} 2$ and the positive rail. Resistor R3 is fixed at $47 \mathrm{k} \Omega$. However, the other base resistor, that of Tr1, is made up from a single fixed resistor R2 ( $12 \mathrm{k} \Omega$ ) and a light dependent resistor, LDR1. An LDR is simply a resistor whose exact ohmic value depends upon the amount of light falling on it. By using this element in the frequency determining section of a multivibrator, the intensity of light striking the LDR controls the frequency of oscillation. A bright light will cause a high note; shadow lowers the note quite considerably.

Transistor $\operatorname{Tr} 3$ acts as a direct-coupled voltage follower. It derives its drive signal from a potential divider formed by Tr2, and its collector load, R4. The remaining components, resistors R1 and LS1 form the loads for Trl and $\operatorname{Tr} 3$ respectively.

## Assembly and testing

Having plugged in all the components as per Fig. 1, check the assembled S-Dec against the component layout shown in Fig. 2. If all is in order, connect a 3 V battery to S -Dec hole 40 (positive) and hole 70 (negative). A tone should be heard. Waving a hand in front of the LDR should vary the pitch of the note. Although a 3 V battery was used with the prototype, the circuit will function perfectly from only 1.5 V . On test, the prototype even worked at 0.5 V ! With 3 V applied, the project draws some 15 mA with the LDR pointed directly at bright
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light. In shadow, the current drops to around 1.5 mA . Using 1.5 V the 'bright light' current was measured as 6.5 mA ; in shadow it went down to 0.8 mA -thus the battery should last a long time.
the seeker (blindfolded) wanders about with the LDR searching for the paper. Use dark paper if the carpet is light coloured. Continuing on this theme, a maze may be laid out and altered for each contestant.

Fig. 2: Plan view of the assembled 5 DeC board. For more permanent construction, the components may be soldered directly on to Blobboard or Veroboard.


The assembled unit. The LDR is shown mounted in a plastic film can to give this component a narrower field of view. As a further refinement, a small lens may be fixed into the end cap with the LDR in the focal plane. This enables a high degree of differentiation between light and dark surfaces-even at a considerable distance.

## Applications

The first application is that of a light-controlled mini-organ. By moving a light closer to or further from the LDR a tune can be played.

It is instructive to play "blind mans buff." By wearing a blindfold, it is possible to find the way around a room because the LDR, when pointed at different things, gives a different tone depending on the reflectivity of the object.

Treasure hunts might consist of pieces of white paper put on the floor. These are the treasure, and

In a more serious application, it could form a guard alarm. The trick here is to point the LDR at a light and adjust R2 until the note goes jusi above audibility. Anyone moving between LDR and light source will cause an audible note thus betraying their presence. Try using a $10 \mathrm{k} \Omega$ potentiometer for R2.
The LDR will function much better if it is shielded in a container of some sort, such as a cigar tube or plastic film can. This prevents extraneous light from interfering with the pitch of the note.


by Eric Dowdeswell G4AR

GOOD news from Martin Sole, long time correspondent of this column. He passed his RAE and is now G8NAT and active on 2 m with SSB using a Liner 2 feeding an 8 -element yagi beam. Martin is going on to get his code test and a 'full ticket' as he puts it. Congratulations Martin, and may you have many happy years on the air.
My request for info on the Crystal Calibrator No. 10 brought forth useful data from three readers, J. $\mathbf{P}$. Bell of London N22, old friend $\mathbf{E}$. Kendall G3APA and C. Pettifer of Slough, who tells me of some interesting ' DX ' he worked with the old Army set 62 for which the calibrator was intended. Cliff very kindly offers to help anyone needing info on old Army sets. He can be found at 90 Uxbridge Road, Slough, Berks. But don't forget, lads, send return postage, a word of thanks and prompt return of anything loaned from Cliff. G3APA mentioned that SWM for Jan 1960 carried some info on mods to the filament circuit of the calibrator.
Michael Walker of Woodlesford, Leeds visiting RAF Locking was delighted to find G3RAF active and to listen to a few QSO's on 2m. Michael gets going on his studies for the RAE very soon so we wish you all the best OM. I was delighted to have a call at home from Geoff Cole G4EMN who had some business in Ashtead. He is secretary of the Wessex AR Group, mentioned before, and one forthcoming attraction is a visit to the School of Signals at Blanford on Sunday 24 July. Club net on 28.575 MHz at 1030 Sundays or write to Geoff at 6 St. Anthonys Road, Bournemouth for more info, or ring 0202 20027.

Derek Gilbert and Ian Horwill share a receiver in Farnham, Surrey using a Heathkit SW717 and a long wire. The reply to queries on several prefixes must be my standard one! Send 35p to Geoff Watts 62 Belmore Road, Norwich for his invaluable prefix list. Cuts out all the guesswork and ensures that your prefix and country count is up-to-date and accurate. D. F. Mathews of 21 Portman Avenue, East Sheen, London SW 14 is very keen on direct conversion receivers and would appreciate any information on articles that have appeared on the subject. Some advance news of the North West Amateur Radio Convention to be run by the Lancaster ARS on Sept 17 and 18 next. Full info and booking forms from J. R. Morris Dept. of Physics, University of Lancaster, Bailrigg, Lancaster.
A first letter from M. G. Hayman BRS 37632 of

London SE2 tells of his interest in SW listening and his intention to take the RAE this year. He started off with a HAC DX Mk2 set but now has a National NCl 20 with an ATU and 66 ft inverted-L aerial. He would like to beg, borrow or buy any info or a manual on the NC120 so if you can help write to 'Woodleigh', 66 New Road, Abbey Wood, London SE2. Robin Bayley A9203 up in Shifnal, Shropshire is still logging the DX on his Eddystone EC10 and long wire, including A2, A6, and AP2 on Top Band! Bob Firth G3WWF got lumbered again with the job of scribe of the Wakefield and District ARS but I'm sure he enjoys his work really! A note to him at 6 Eastfield Drive, Woodlesford, Leeds or a ring to Leeds 825519 will bring all the gen on the society's activities. On 19 July G4DZU will talk on moonbounce and meteor scatter techniques while G3MFJ and Bob will be giving a lecture and demonstration of RTTY on 16 August.
Regular Brian Harrison has found the 10 m band open down in Ore, Hastings with PY, CP, HC, PJ and ZP representing S. America. On 15 m ZS3, CP, P2 (Surinam) and 9 Vl were logged. Old friend Alan Doherty BRS34968, Portrush, Co. Antrim seems to have concentrated on the Pacific on 20 m with 'dozens of KL7, KH6 and VE8! Alan operates the local marine VHF station around 156 MHz and uses the same aerial system for 2 m ! Oscar 7 provided KL7, VE2 and W9 on the 2 m downlink. Not bad when one remembers that the up link is on 432 MHz ! At the other extreme Alan now has 208 countries on 80 m the latest being XT3, YS, A9, LU and TI.
In Cambridge Dave Peck BRS37621 reports a new type of QRM but this time he is causing it! Neighbours report QRM from the governor contacts on his RTTY machine! But this did not stop him from copying RTTY from A4, EA3, JA3, LA, OH, W and 9M2. Dave is starting up on 2 m and is looking for a yagi to go with his rotator, cheap of course, and at least 10 elements. Simon Robinson in Stocksfield, Northumberland, kindly sent me an old 'giveaway' dated October 1924 from Popular Wireless, which described how to make five 'special' crystal sets! Simon bought a pre-war Pilot radio set which only needed a couple of new electrolytics to get it going again. At 20 p it was a cheap way to get on the SW bands! Now he has an ex-Army set and has given the Pilot away to an elderly invalid. But he still would like a proper communications receiver!

It was good to hear from John Hodgson (Morpeth) again after a long interval. He has progressed from a Mini-Clipper to a Realistic DX160 plus a 40 ft long wire although John stuck to 20 m and seems to have ignored the bandswitch! Pity! The log from D. Waddell located in Herne Bank was a joy to my eyes since the majority of the entries were for CW stations on 15 and 20 m . In one period, 1145 to 1345 , no less than 29 Japanese stations were copied on 15 m CW. I'd like to hear from more of you who can copy CW and don't forget that there is often DX to be heard on CW when the SSB parts of a band seem to be dead.

## Log Extracts

D. Waddell:-20m CP3CN FM7WV F08EI FY7AS HC1LT HK0BKX HK7UL JRGRRD (Okinawa) YSIO 9M8HF CP6EL FH8DF HH2MC TU2EF VS5MC 5U7AG 15m CE8AA CP6FN FG7AM TAIZB UM8AX CP6EL TR8SM 5N2NAS 5X5NK

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All SSB (or RTTY) except those in bold which are CW.


## MEDIUM WAVE DX

## by Charles Molloy

"OW does one QSL a broadcasting station" writes Andrew Grendon from Dun Laoghaire in the Irish Republic. The procedure, very briefly, is to send the station a reception report and hope that this will lead to a QSL card or verification letter in return. International broadcasters on the short waves are very pleased to receive letters from their listeners and provide colourful and interesting QSL cards and pennants for them. On the medium waves the DXer is usually outside the service area of his quarry and he really has the status of an eavesdropper! A lot more care then is required when writing to a medium wave station as one is depending on goodwill for a reply. Return postage in the form of an International Reply Coupon, obtainable from main post offices, should always be enclosed.

The object of the reception report, which is sent to the Chief Engineer, is to convince him that the DXer really did hear his station. Programme details, weather reports, commercials (jingles); station identification slogans, names of announcers, time checks, items of local news-the writer was able to give the name of the winner of a local beauty competition when writing to Radio ZDK in Antigua 1100 kHz these are the sort of items that should go into the report. The date, the time in either GMT or local time and of course the frequency in kHz should be mentioned. The report should end with a request for a confirmation of reception. The best time for obtaining programme information is during the period five minutes before the hour until five minutes after the hour, when programmes change and station identifications and commercials are broadcast.
"North American stations are good verifiers," writes Ian Rennison from Horsham in West Sussex. He mentions CKBW in Bridgewater Nova Scotia on 1000 kHz whose Chief Engineer sent him an attractive QSL card, a two page sheet giving the history of the station, a paper mat printed in colour showing the coverage of CKBW, a leaflet giving a description of Bridgewater, a road map of the town and some tourist information of the area. Whew, all he wanted was a verie! Ian uses a Trio 9R59D receiver and a medium wave loop with differential amplifier. His $\log$ which covers the winter period includes 44 North American outlets plus Radio Belize on 834 kHz ; Radio Antilles, Montserrat on 930; Radio Jornal do Brasil on 940; Radio Margarita, Venezuela on 1020 and an unidentified broadcast heard at sunset on 1540 kHz . The latter might be the Voice of Peace located in the Red Sea area and reported by a DXer in Cape Town in the June issue.

It is not essential to know the postal address when writing to stations in North America. "The Chief Engineer, WINS Radio, New York City, NY, USA" is typical of the format to use and station callsignis and the name of the town or city are broadcast frequently. When writing to other parts of the world is is usually better to use the postal address which can be found in the World Radio and TV Handbook, which comes out annually and can be ordered from bookshops.

Lawrence Bennett of Bristol is keen on obtaining QSL's. He reports receiving a QSL card from WINS on 1010 and WNEW on 1130. He says "I have been trying my hand at writing letters to stations in foreign languages and have had rather a good response." Radio Espana de Madrid, EAJ2 on 917 kHz sent him a beautiful QSL card for a report in Spanish and Sud Radio Andorra 818 kHz replied to a report in French with a card and two stickers. No need to be a linguist to write a report in a foreign language. Many DX clubs provide their members with report forms and vocabularies in various languages and all the DXer has to do is to fill-in the spaces with the appropriate details. A few years ago a German club produced a report form in Arabic which really was an achievement!
Tuning across the band, looking for interesting signals is the usual procedure adopted by DXers and a very good method it is too. New and unexpected broadcasts are heard this way. A few DXers tune to spot frequencies just to check on conditions. If CJON cannot be heard on 930 then it is unlikely that any other North American will be heard. There is a third, less known method of DXing and that is to tune to a chosen channel and remain on it waiting to see what comes up. Obviously the frequency should be selected with care.

Two interesting channels are 1484 kHz and 1594 kHz which, in the European area, are known as the International Common Frequencies. Large numbers of low power locals transmit here and in some parts of the UK they are used by local radio. On 1484 there are the American Forces (AFRTS) broadcasts from Iceland and Kenitra in Morocco as well as CSB90 in Madeira and Radio Gibraltar while 1594 has AFRTS in Athens, Manx Radio, Coimbra in Portugal and many others. In North America 1230, 1240, 1340, 1400 and 1490 are known among DXers as the Graveyard Channels and are the resting place of hundreds of local stations in the United States plus ZBM1 in Bermuda on 1230 and ZBM2 on 1340.

Each DXer will find his own favourite channels. For a start try 960 kHz which is the home of CHNS in Halifax Nova Scotia, Radio Nacional Mendoza in Argentina, Radio Sutatenza in Colombia, Radio Victoria in Aruba (Netherland Antilles) which has programmes in English, and Music Radio in Hamilton Bermuda.
Radio Finland is now broadcasting its external service on two frequencies on the medium waves, 557 kHz and 962 kHz . The English programme which includes an item for DXers is on Thursdays and Saturdays at 2030 GMT .


## SHORT WAVE BROADCASTS by Derek Bell

THERE is a famous poem that thinks on things gone by and asks if there are still crumpets for tea, or words to that effect! A letter arrived chez Bell recently that asked if the Derby area had changed much. The writer Harold Bailey, now living in Cornwall, retired in 1965 after a lifetime's work in this area. Harold told tales of the old days of short wave DXing in the 1920's when "valves glowed like Blackpool illuminations" and one could hear America free of interference in the wee small hours. He also made a 6 -inch TV set after Sutton Coldfield opened and was a member of the Derby local amateur radio society when it was known as the Derby and District 1911 club.

John Cook from Southend is also a long-time DX fan. He has reached the lofty heights of possessing a Grundig Sattelit and below is a handful of his loggings:-

4980 Radio Ghana at 2230
4990 Radio Nigeria at 2230
11765 FEBC at 0830
11835 4EVH at 2330
15335 Rad. Nac. de Colombia at 2235
15425 Radio Sri Lanka at 1230
We now devote a short time to QSL fans thanks to the good offices of Roy Patrick from Derby. He supplies this month's station information on WYFR the American Gospel station. They are building a new transmitter site 16 miles from Lake Okeechobee, Florida. The Scituate site will remain on the air until Florida is ready. They are also quoting a new address namely PO Box 157, London NW1, for QSL loggings. The other famous Gospel station HCJB, Quito, Ecuador has issued a new set of QSL cards showing transmitter equipment. Radio Canada have sent me their new card, in two shades of blue and white and it shows three examples of the Canadian communications satellites. These are the "Hermes", "Alouette" and "Anik" and along with the card goes a leaflet giving a brief technical description of each. This is another card that follows the RCI practice of having to be sent back to them for verification.
"What does 'DX' mean?" asks R. J. Irvine of Selly Oak. This is a question often asked, and the answer
is that as initials they mean nothing. They are two letters chosen to stand for long-distance reception. In other words Radio New Zealand is DX reception for us in the UK compared to, say, Radio Nederland but as with many other initials they have tended to be applied to any distant station. (How about "Distance Xtra"? A.E.D.) R. J. also tempts me with the news that he has lashed up a SW set with two BC108's a 365 pF varicap (?) and a coil. It sounds very simple so may I ask you to send me the diagram R.J. and then we can see if your loggings are justified.
The one great help that any DXer can have is a club. A friend of mine is convinced that DXers are hermits by nature but I have had a letter asking if I could mention a rapidly growing club namely the Twickenham DX Club, 37A Popes Grove, Twickenham, Middx., so with my compliments Gentlemen let's hope that the Hermit tag is a false one! Mr. C. Slow asks if he is allowed to submit reports to PW since his set is only a HMV domestic set with a 90 ft aerial in the loft. This column, as I have mentioned before, is written by you, the readers, and I merely pass on the news and views of your good selves to others having the same interest. Many readers I am sure feel that they are not true DXers since they only possess domestic or portable sets. Many is the reader I have had send logs in from their home sets and then go on to be deeply involved and in some cases go on to the ultimate and take the RAE exam. Mr. Slow has some complimentary things to say about this column helping DXers for which my thanks.
Finally the latest edition of the Radio Berlin International newspaper has arrived and tucked away among the reports is their latest schedule. This shows that English is transmitted on 6080, 6115, 7185, 7300, 9730 and 7260 at 1830, 1930 and 2215, and with that I will' close wishing you and yours best 73s.


IT shows the extent of amateur radio when a specialist column like this can report about readers activities from microwavelengths of 3 cm to the relatively long wavelengths of 10 m .
Many visitors to the RSGB exhibition at Alexandra Palace (May 6/8) were attracted to the lectures and demonstrations given by the Microwave Group, especially the reception of live TV pictures on 3 cm $(10 \mathrm{GHz})$ from Ilford, 14 km away. Peter Kerry G8ARO Farnham, tells me that "almost daily" he and his fellow enthusiasts hear of amateurs who are building gear for 10 GHz and he is worried in case some give up in despair through the lack of a bit of advice or a trial QSO. He has asked me to say that any of the Southern Group would always seek to help a newcomer to microwaves. On April 24 G8ARO had a 59 contact with G3BNL/P over 60 km between Hindhead and Old Reading on the GLC/ Herts border, and again from Hindhead, on May 1st, he worked G8GKV/P and G3JHM/P, both 59, situated on Chanctonbury Ring, Sussex, a distance of

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| No. | (Watts) | $£$ | P\&P |
| 07 | 20 | 4.40 | .79 |
| 149 | 60 | 6.20 | .96 |
| 150 | 100 | 7.13 | 1.14 |
| 151 | 200 | 11.16 | 1.50 |
| 152 | 250 | 12.79 | 1.84 |
| 153 | 350 | 16.28 | 1.84 |
| 154 | 500 | 19.15 | 2.15 |
| 155 | 750 | 29.06 | OA |
| 156 | 1000 | 37.20 | OA |
| 157 | 1500 | 45.60 | OA |
| 158 | 2000 | 54.80 | OA |
| 159 | 3000 | 79.05 | OA | Primary 220-240 Volts


| Ref. | Amps |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| No. | 12 v | 24 y | £ | P\&P |
| 111 | 0.5 | 0.25 | $2 \cdot 20$ | - 45 |
| 213 | 1.0 | 0.5 | 2.64 | . 78 |
| 71 | 2 | 1 | 3.41 | 78 |
| 18 | 4 | 2 | $4 \cdot 03$ | 96 |
| 70 | 6 | 3 | $5 \cdot 35$ | . 96 |
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Ref.
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112
79
3
20
21
51
117
88
89

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200 238200 A-0-3 $\begin{array}{llll}212 & 1 \mathrm{~A}, 1 \mathrm{~A} & 0-6,0-6 \\ 13 & 100 & 9-0.9\end{array}$ $\begin{array}{llll}135 & 100,330 & 9-0.9 \\ 230 & 0-9, & 0-9\end{array}$ $\begin{array}{lll}235 & 330,330 & 0-9,0-9 \\ 207 & 500,500 & 0-8-9, \\ 0 & 0-8-9\end{array}$

$$
\begin{aligned}
& \text { Prim. } 200 / 220 \mathrm{~V} \text { or } 400 / 440 \mathrm{~V} \\
& \text { Sec. } 100 / 120 \mathrm{~V} 200 / 240 \mathrm{~V}
\end{aligned}
$$ 208 1A, 1A 0 0-8-9, 0-8-9 236 200, 200 0-15, 0-15 $214300,300 \quad 0-20,0-20$ 221 700 (d.c.) 20-12-0-12-20 2061 A. 1 A $\quad 0-15-20-0-15-20$ 203 500, 500 0-15-27-0-15-27 $2041 \mathrm{~A}, 1 \mathrm{~A} \quad 0-15-27-0-15-27$

\% Ref
No
124
126
127
125
123
40
120 HIGH VOLTAGE MAINS AO

$$
\begin{aligned}
& \text { ISOLATING } \\
& \text { PrIm. 200/220V or } 400 / 440 \mathrm{~V}
\end{aligned}
$$

$$
\begin{array}{rrrr}
\text { Sec. } 100 / 120 \text { V } & \text { or } 200 / 240 \mathrm{~V} & \text { P \& } F \\
\text { VA } & \text { Ref. } & 5 & 5.89 \\
30 & 243 & 1.32 \\
350 & 247 & 14.11 & 1.84 \\
1000 & 250 & 35.65 & \text { AO } \\
2000 & 252 & 54.25 & \text { AO }
\end{array}
$$

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| :---: | :---: | :---: |
| 400 V | 2 A | 0.55 |
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| 119 |
| 60 |
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34 km . In four months operating, over a winter period, Peter has a 10 GHz county score of six, from his regular site at Hindhead.

At 1908 on May lst, Nigel Golds, BRS36910, West Chiltington, Sussex, received a 599 signal from $5 B 4 C Y$ on 10 m , and I received strong signals from the German beacon, DLOIGI, at 1453 on the l1th, 1815 on 16 th, 0815 on the 17 th and 1345 on the 20 th. The 10 m band was open to CT, UK, YU, DJ and OK during the late aficmoons of the 11th and 16th. Short skip conditions on 10 m can often be linked with seasonal sporadic-E.

The first real opening of the 1977 sporadic-E "season" occured during the early morning of the 4th of May when $59+$ signals poured into Sussex from about nine east-European broadcast stations, operating between 68 and 70 MHz . On most days since late April there have been prolonged bursts of signals on the Rl vision channel, $49 \cdot 75 \mathrm{MHz}$, often accompanied by simultaneous bursts from Polish FM stations in the 4 m band. Another region of the spectrum vunerable to sporadic-E disturbances is the $6 / 7 \mathrm{~m}$ band, and, during the early mornings of the 4th, 11th, 16th and 17th May, many continental radio telephone signals were heard between 40 and 46 MHz .

## BROADCAST BANDS

Short Wave Reports by the 15th of the month to Derek Bell, 169 Max Rd., Chaddesden, Derby. Medium Wave L'ogs to Charles Molloy, 132 Segars Lane, Southport, PR8 3JG.

## AMATEUR BANDS

Logs covering any amateur band/s in band/ alphabetical order by the 25th of the month to Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey, KT21 2TW.

## VHF

Reports on VHF matters to Ron Ham, Faraday, Greyfriars, Storrington, Sussex RH20 4HE.

Frank Luman, Glasgow, a Band II enthusiast, frequently heard Downtown Radio from Belfast on 96 MHz during the first few days in May, which to him, is a good indicator of improved conditions. Frank has tried a variety of VHF arrays and is now considering giving a crossed yagi a try to improve his reception of those stations with circular polarisation. Keep us informed Frank, aerials are a most interesting subject.

Our regular contributor, Alan Baker G8LGQ Newhaven, who, incidently has been elected Chairman of the newly formed Brighton and District Radio Society, worked F, ON, PA and DJ in the 144 MHz section of the IARU Region 1 contest on the 8 th May. The best DX heard was a DM at 500 miles; although there was no indication of a tropo opening, Alan found the conditions most unusual. Both the Wrotham and French 2 m beacons were below normal with him, yet there were frequent bursts of signals lasting about half a minute from the DX stations. Both Alan and myself monitored two brief tropospheric openings the first from late on the 17 th to the early hours of the 18th and again from late on the 19 th to early on the 20 th . While our respective QTHs were lịnked by landline we heard a G station
work LA6HL, strong signals from GW and from London stations who were beaming north. On the second occasion, we both heard strong signals from ON5NY, GW and northern G. Throughout each event I received a 569 signal from GB3EM on 70 cms with only a dipole feeding my receiver.

Cmdr Henry Hatfield, Sevenoaks, observed a sunspot group and a plage on the 6th May and a small solar flare at 1027 on the 18th, with his spectrohelioscope. Both Henry and myself recorded bursts of solar radio noise at 136 MHz on the 8th and G8LGQ heard the noise at 144 MHz . There was an auroral alert during the evening of the 11th and Charlie Newton G2FKZ, London, said that some northern VHF signals were taking on an auroral tone. What about it readers, did you hear anything unusual that evening? G2FKZ, who is the RSGB auroral information co-ordinator, tells me that during the second phase of an aurora from about 2300 on May 2nd to 0130 on the 3rd, GM operators worked some 70 stations spread through 20 countries, including OH . Recent auroral activity reports, especially those containing beam headings, have made it possible for G2FKZ to establish a more definite auroral pattern.

From down under, another welcome letter from Anthony Mann, Applecross, W. Australia who says that conditions during April became unusually active. During the evening of the 11th a VHF transequatorial opening occurred and he received signals on R1 from two Chinese stations until about 1930 (local time). He was alerted to this opening at 1900 when there was a lot of activity from JA between 27 and 32 MHz . On the 16 th Anthony heard a SouthKorean (KBS) station on $44 \cdot 3 \mathrm{MHz}$ during the afternoon along with Russian stations on 38 and 44 MHz . 10 ml was particularly active on the 11 th and 16 th and the International beacons 5B4CY, DLOIGI, and 3B8MS were heard around 1610 on the 6th. There was intense 10 m activity on the 17 th , but the MUF did not rise above 35 MHz . Some sporadic-E disturbance was observed during the afternoon by a TV. DXer in NSW.

Thanks again for your most interesting reports. and believe me, they really are valuable.

## HIDLDNDI: <br> A 741 Signal Tracer July '77

In Fig. 1 and 4 one of the diodes D1/2 should be reversed to ensure that they are back-to-back.

## Tele-Games June 1977

The mounting of the voltage regulator 1 C 2 on the PCB, Fig. 6, is not too clear. The IC is mounted with the cooling fin uppermost, away from the PCB. The heatsink is made from a small piece of aluminium, bent at right angles, and bolted to the fin. See photograph on page 130.


THE success of the RSGB International Radio Communication Exhibition And Convention held at Alexandra Palace, London, over the week-end May 6-8, was obvious by the fact that some 6,000 people attended it. Both the Society, and John Hitchins G4FGN, who masterminded the whole affair, were justly rewarded by the enthusiasm of the visitors as they wandered around the 30 or so trade stands in the Grand Hall. The home constructor was especially catered for; he could buy anything from a host of small components, at reasonable prices, to a pneumatic mast and a super VHF array and rotor, to put on the top. In addition to the ever popular surplus gear, the traders were offering a wide range of the latest converters, receivers, transmitters and test gear, and, that all-important friendly bit of advice for their prospective customers.
There was something for everyone, films on Friday by ARRL, EMI-Varian, and RSGB, lectures on Saturday and Sunday on Aerials, Microwaves, Radio Astronomy, Raynet, RTTY, Synthesizers, the SWL,
and VHF. Visitors could chat with HQ staff on the RSGB's own stand, visit the Society's book stall, or see the exhibitions by the Amateur Radio Mobile Society, British Amateur Television Group, RAIBC, RAYNET, AMSAT-UK, the Microwave and Propagation groups, and the Royal Corps of Signals.

At the dinner on Saturday evening, Mr R. W. Cannon, Technical Director Cable and Wireless, proposing the toast to the Society, referred to amateur radio as a national asset and said that he admired the way in which the "trade" had produced some very technical equipment at prices which most amateurs could afford. Replying, Dr John Allaway G3FKM Immediate Past President, thanked Cable and Wireless for supplying the regular HF propagation predictions for the Society's journal and he stressed the need for the RSGB to work with other national societies to persuade the newer countries to accept amateur radio. There was spontaneous applause from the 180 people present when Dr Allaway congratulated the Society's President, Lord Wallace of Coslany, on his appointment as the first Lord-in-Waiting to Her Majesty the Queen.

Proposing the toast to the Guests, among whom were representatives from France, Belgium, Yugoslavia and the-USA, Dr Dain Evans G3RPE Executive Vice-President, spoke of the many sides to amateur radio and referred to Ed Tilton, WIHDQ, who gave one of the afternoon lectures, as "a legend in amateur radio". Replying for the guests, Dr J. A. Saxton, Director of the Appleton Laboratory, urged every radio amateur in the United Kingdom to give his wholehearted support to the RSGB and went on to congratulate all those responsible for this event at Alexandra Palace.

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    - Compact $4^{\prime \prime} \times 4^{\prime \prime} \times 4^{\prime \prime}$

    PRO-STROBE $£ 32.50$
    6-8 Joules
    External trigger

    - Long Llfe tube timer
    clrcult
    


    ## 150 WATT LIQUID

    WHEEL PROJECTOR- Accepts al| accessorlos
    - C/w with wheel \& motor plate
    Sturdy teel construction Remarkable value-
    £39.50. Our price
    Is only: $\quad £ 33.00$

[^4]:    / 1 To be added
    

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