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Easytune MW-LW Receiver Practical Audio Amplifiers

## RADIO EXCHANGE LTD.

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## NEW EDU-KIT MAJOR Completely Solderless Electronic Construction Kit. Build these projects without soldering iron or solder

soldering iron or solder



Total Building Costs **£9-00** 

P & P + Ins. 85p

- ★ 4 Transistor Earpiece Radio Signal Tracer Signal Injector
- Transistor Tester NPN-PNP
- \* 4 Transistor Push Pull \*
- Amplifier 5 Transist Transistor Push Pull Amplifier
- ★ 7 Transistor Loudspeaker Radio MW/LW
- ★ 5 Transistor Short Wave Radio
- \* Electronic Metronome
- \* Electronic Noise Generator
- \* Batteryless Crystal Radio
- One Transistor Radio 2 Transistor Regenerative Radio
- 3 Transistor Regenerative Radio
- Audible Continuity Tester
- Sensitive Pre-Amplifier

Components include: 24 Resistors 
\$\pm\$ 21 Capacitors 
\$\pm\$ 3" \times 3" Loudspeaker 
\$\pm\$ Earpiece 
\$\pm\$ Mica Baseboard 
\$\pm\$ 3 12-way connectors 
\$\pm\$ 2 Volume controls 
\$\pm\$ 2 Slider Switches 
\$\pm\$ 1 Tuning Condenser 
\$\pm\$ 3 Knobs 
\$\pm\$ Ready Wound MW/LW/SW Coils 
\$\pm\$ Ferrite Rod 
\$\pm\$ 6y yards of wire 
\$\pm\$ 1 Yard of sleeving, etc. Complete kit of parts including construction plans.

#### VHF AIR **CONVERTER KIT**

Build this Converter Kit and receive the Aircraft Band by placing it by the side of a radio tuned to Medium Wave or the Long Wave Band and operating as shown in the Instructions supplied free with all parts. Uses a retractable chrome plated telescopic aerial, Gain Control, V.H.F. Tuning Capacitor, Transistor,

All Parts including Case and £3.95 Plans. P & P and Ins. 45p



WITH Multiband V.H.F. section covering Mobiles, Aircraft, T.V. Sound, Public Service Band, Local V.H.F. Stations, etc. and Multi-band A.M. section with Airspaced Slow Motion Drive Tuning Capacitor for easier and accurate tuning, covering M.W.1, M.W.2. L.W., Three Short Wave Bands, S.W.1, S.W.2, S.W.3. & Trawler Band, Built in Ferrite Rod. Aerial for Medium Wave, Long Wave and

Trawler Band, etc., Chrome Plated 7 section Telescopic Aerial, sngled and rotatable for peak Short Wave and V.H.F. reception. Push Pull output using 600mw Translstors. Gain, Wave-Change and Tone Controls. Plus two Slider Switches. Negative Feedback circuit and SPECIAL POWER BOOSTER SOCKET AND virtually double gain if required. RESISTOR, to virtually double Powered by P.P.9 9 volt battery.

Complete kit of parts including carrying strap. Building instructions and Operating Manuals.

£13.99 Including Post and Packing Case Enclosure Kit (if required) £1:80 inc. P & P.



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Total Builds £3.60



ECK2 Self Contained Multi-Band sistors and 3 Diodes Push/Pull output. 3" Loudspeaker, Gain Control. Superb 9 section swivel ratchet and retractable chrome plated telescopic aerial, V.H.F. Tuning Capacitor, Resistors, Capacitors, Transistors, etc. Will receive T.V. Sound, Public Service Band, Aircraft, V.H.F. Local Stations, etc. Operates from a 9 Volt PP7 Battery (not supplied with kit). Complete kit of parts including construction plans

£7.95 P & P and Ins. 70p



FUSA 7 Transistors, 6 tuneable wavebands, M.W., L.W., Trawler Band, 3 Short Wave Bands. Receiver Kit. With 5" × 3" Loudspeaker. Switch. 7 Transistors and 4 Dioles. 6 section chrome-plated telescopic aerial. 8" Sensitive Ready Wound Ferrite Rod Aerial. Tuning Capacitor. Resistors, Capacitors, etc. Operates from a 9 Volt PPT Battery (not supplied with kit). Complete kit of parks including construction plans. 6.7 RE

construction plans. £7.25 P&P and Ins. 70p.



### **EDU-KIT JUNIOR**

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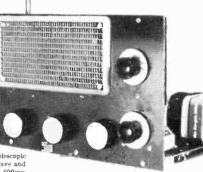
Radio

3 Transistor Earpiece Radio Medium Wave Coverage.
4 Transistor Medium Wave Loudspeaker Radio.

\* Electronic Noise Generator

All parts including Loud-speaker, Earpiece, MW Ferrite Rod Aerial, Capacitors, Re-sistors, Transistors, etc. Complete kit of parts includ-ing construction plans.

£ $6.55_{P\&P+Ins.60p}$ 



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Total Building costs £4.30 P&P + Ins. 60p.

E.V. 6. Case and looks as above. 6 Transistors and 3 diodes. Powered by 9 voit battery. Ferrite rod aerial, 3" loudspeaker, etc., MW/LW goverage. Push Pull output. All parts including Case and Plans.

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E.V. 7. Case and looks as above, 7 Transistors and 3 diodes. Six wavebands, MW/LW, Trawler Band, SW1, SW2, SW3, powered by 9 volt battery. Push Pull output. Telescopic aerial for short waves. 3"

All parts including Case and Plans

Total Building costs £6.95 P & P + Ins. 70p.

To RADIO EXCHANGE LTD	
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Tel: 0234 52367	

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# PRACTICAL WIRELESS

VOL. 52 NO. 11 ISSUE 841 MARCH 1977

BRITAIN'S PREMIER MAGAZINE FOR THE DO-IT-YOURSELF RADIO AND ELECTRONICS CONSTRUCTOR

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FREE THIS MONTH!
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—See page 938 for full details

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## Can YOUR Antenna do all this?

A SMALL SELECTION FROM OUR HUGE FILE OF TESTIMONIAL LETTERS ON THE JOYSTICK VARIABLE FREQUENCY ANTENNA ( 5-30MHz)

G4DJY's COPY LOG shows 68 North Americans worked in the ARRL W/VE contest.

W6TYP worked WA6JPR over hundreds of miles on 40m. using the JOYSTICK VFA and MICRO-WATTS "equivalent to 1,000,000 MILES PER WATT"-A WORLD RECORD-we can supply conclusive evidence!

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W7OE U.S.A. Government Electronics Engr. (retd.), claims "VFA 5ft below ground, same as dipole, elevated 15ft, one "S" point UP on dipole".

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128 2	θp	117	18p*	301	34p	BFY50	25p	ZTX300	13p*	
151 3	5p	142	24p	303	35p	51	25p	302	18p*	
153 3	5p	143	24p	338	16p*	52	25p	500	15p*	
176 2	2p	147A	9p*	557	15p*	BY164	40p	502	18p*	
187 2	2p	147B	10p*	BCY71	18p	ME0401	18p*	1N914	5p	
188 2	θp	148	9p*	72	14p	0411	18p*	1 N4001	\$p	
AD149 6	₽p	148B	10p*	BD123	90p	ME4001	10p*	4002	6p	
161 5	2p	149B	11p*	124	90p	4101	11p*	4005	\$p	
162 5	2p	149C	11p*	131	42p	MPF102	40p*	4006	10p	
		154	18p*	132	42p	OA47	14p	4148	5p	
MCH/PR 1		157B	12p*	139	54p*	90	7p	2N2219	30p	
AF116 2	4p	177	17p	155	75p*	91	₽p.	2646	45p	
117 2	4p	178B	18p	BF158	20p*	200	10p	2926O	13p*	
124 3	0p	179B	19p	184	25p	OC28	1 . 20	2926G	15p*	
	6p	184B	12p*	185	28p		1 - 20	3053	25p	
AU113 1-	72*	184L	11p*	194	10p*		1 . 20	3055	60p	
BC107 11	1p	187	26p	196	12p*	44	45p	3702	11p*	
107A 1	2p	212A	13p*	197	12p*	71	35p	3703	14p	
108 10	0p	212L	15p*	199	15p*	ORP12	68p	3704	13p*	
L CIL108 (	6p	2138	12p*	200	38p	TIP29A	47p*	3705	14p	
BC108B 11	1p	214	15p*	338	34p	30A	56p *	3707	12p*	
108C 12	2p	214L	17p*	BFW11	68 p	31 A	57n	3819E	25n*	
DIODES		BRIDGE		RESIST	OPE	IInt anah	LINE	AR I/C'	s "	
	130	100V 1A		WATT	E12 /	ogp oacii	741		33p	
		200V 1A	30n	1 ohm - 1	Om oh	'/o'	LM38	0 SL60745	1-29*	
		400V 1A	32p				LM39		68p	
	npl			ZENERS	(400m	W) BZX 83			60p	
		_	- 1	3V, 3V3,	5V1,	5V6, /V5,	SN76	013ND	1-60*	
THYRIST	5		9V1, 10V,	127	184, 224,	SN76	013N	1 - 75 *		
60V 1 A	20	P	- 1	30 V. AII	at 12	ртеасп.	SN76	023N	1 - 75 *	
600V 1A	60	TAG 1		LED Til	209/0	·125 0·2"	TBA	800	1 - 35 *	
700V 1 A		BT106	- 1	Red	20p	20p	TBA	810S	1 - 49*	
400V 4A		C106D1		Grn	29p	29p	TBA	320	1 - 20 *	
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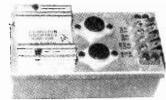
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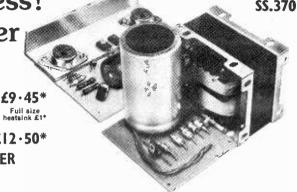
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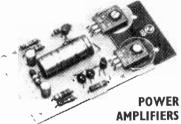
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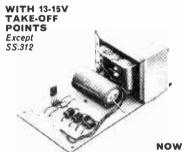
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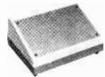
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1												
ł	OA2	-85	6AX4	-75	6JU8A	-90	12BE6	-55	:30PL1	1.00	DF96	-66
1	OB2	-40	6B8G	-35	6K7G	-35	12BH7	-55	35A3	-75	DK40	-70
ı	OZ4	-55	6BA6		6K8G	-50	12BY7	-85	35C5	-80	DK92	1.00
ı	1A3	-60	6BC8	-90	6K8GT	-55	12E1	3.50	35 D5	-90	DK96	-70
ı	1A5GT	-55	6BE6	-40	6L1	2.50	12J5G7	Γ -40	35 L6G	r -80	DL96	-60
ı	1A7GT	-60	6BH6	-70	6L6GC	-70	12J7G7	.70	35W4	-55	DM70	-80
ı	1B3GT	-55	6BJ6	-85	6L7	-60	12K5	1.50	35 Z3	-80	DM71	1.78
ı	1H5GT	-80	6BK7A	-85	6L18	-60	12K7G	T -50	35Z4G7	r ·70	DY87/6	-85
1	1L4	-25	6BQ7A	-60	6L19	2.00	12K8	-75	35Z5G7	r ⋅80	DY802	-45
ı	1N5GT	.75	6BR7	1.00	6LD20	-80	12Q7G'		42	1.50	E80CC	2.50
ı	1R5	-50	6BR8	1.25	6N7GT	-70	128C7	-50	43	1.25	E80CF	5.00
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ı	1U4	70	6BW6	1.00	6Q7GT	-50	128H7	-50	50C5	-70	E88CC	1.20
ł	1U5	-85	6BW7	-65	6Q7(M)	-65	128J7	-60	50CD60		E92CC	.70
ı	2D21	-55	6BZ6	-60	6R7G	-70	128Q7	-80	ľ	1.80	E180F	1.15
ı	2GK5	.75	6C4	-40	68A7	-55	128R7	-75	50EH5	-85	E182C0	
П	2X2	-70	6C5G	-60	68C7GT		14H7	-75	50L6G			8-00
П	3A4	-55	6C6	-45	68G7	-50	1487	1.00	85A2	.75	E188CC	,
П	3D6	-40	6C9	2.00	68H7	-55	19AQ5	-65	85A3	-75		2.50
П	3Q4	-80	6CB6A	-50	68J7	-60	1906	6.50	150B2	1.00	E1148	-60
Н	3Q5GT	-70	6CD6G	1.60	68K7G7		19H1	4.00	2158G	-60	EA50	-40
Н	384	-45	6CG8A	-90	68Q7	-60	20D1	-70	807	1.10	EA76	1.30
П	3V4	-80	6CL6	-75	6V6G	-80	20F2	-85	1625	2.50	EABCS	0 -40
ı	4CB6	-75	6CL8A	-95	6V6GT	-55	20L1	1.20	5702	1.80	EAC91	-55
ľ	5CG8	.75	6CM7	1.00	6X4	45	20P1	1.00	5763	1-65	EAF42	-70
ı	5R4GY		6CU5	-90	6X5GT 7B6	45 80	20P3	1.00	6057	1.00	EAF80	75
ı	5U4G	-60	6D3	.75	7B7		20P4	-84	6060	1.00	EB34	-30
ı	5V4G	-60	6DE7	-90	7H7	·80	20P5	1.50	6067	1.00	EB91	-17
1	5Y3GT	-55	6DT6A	-85	7R7	2.00	25 A 6G	.70	7025	1 50	EBC41	-75
ı	5Z3 5Z4G	1.00	6EW6 6E5	1.00	7Y4	-80	25L6G	-70	7193	-60	EBC81	-45
ı	5Z4GT	-55	6F1	1.00	7Z4	-80	25 Y 5 G 25 Z 4 G	-60 -50	9002 9006	-55	EBF80	-40
ı	6/30L2	.79	6F6G	-60	8D2	-50	25Z5	-75		45	EBF83	-45
ı			6F13	-90	9D7	-70	25Z6G	-80	AC6PE		EBF89	-40
ı	6AC7	-55	6F14	-90	10C2	.70	28D7	2.00	AL60	1.20	EBL21	2.00
ı	6AG5		6F15	-85	10DE7	-80	30 A 5	-75	ARP3	-60	EC86	-84
ı	6AG7	-60	6F18	-60	10F1	-67	30C15	-77	ATP4	50	EC88	-84
	6AH6	-70	6F23		10F9	-65	30C17	.77	AZI	-50	EC92	-55
h	6AJ5	-70	6F24		10F18	-65	30C19	1.00	AZ31	-60	ECC33	2.00
ı	6AK5	-45	6F25	1-00	10LD11	-75	30F5	.70	AZ41	-50	ECC35	2.00
ı	6AK6	.70	6F28	-74	10P13	-80	30FL14		B36	-75	ECC40	-90
ı	6AM8A	-70	6F32	-70		2.50	30L15	.75	BL63	2.00	ECC81	-84
	6AN8	-70	6G6G	-60	12A6	-65	30L17	-70	CL33	1.75	ECC82	-84
ı	6AQ8	-47	6GH8A	-80	12AC6	-80	30P4M1		CY1C	1.00	ECC83	-84
ı	6AR5	-80	6GK5	.75	12AD6	-80	30P12	-74	CY31	.70	ECC84	-35
ľ		1-00 j	6GU7	-90	12AE6	-80	30P19/	- 1	D1	-50	ECC85	-89
ı	6AT6	50	6H6GT	-30	12AT6	-45	30P4	-90	DAF91	-35	ECC86	1.25
	6AU6	-40	6J5GT		12AU6	-50	30PL1	1.00	DAF96	-60	ECC88	-51
	6AV6		6J6	-35	12AV6	-60	30PL13		DC90	.70	ECC189	
Ĺ	8AW8A	-941	6J7G	-35	12BA6	-50	30PL14	1.29	DF91	-80	ECC807	1.40
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	APOLOG	00	BULBIOOS	1 0000 .48	1 230 .90		AC177 -33	
	ECF82		2.50	POC189 -52	QQV03/10	U404 -75	ACY17 -30	BY114 -21
	ECF86	-80	EY51 -45	PCF80 -40	2.00	U801 -80	ACY18 -22	
	ECH21	2.00	EY81 -45	PCF82 -45	Q895/10	U4020 -75	ACY19 -22	
	ECH35	1.60	EY83 -60	PCF84 -70	1.00		ACY20 .21	
	ECH 42		BY84 1.20		Q8150/15	VP41 -90	ACY21 -22	
	ECH81		EY86/7 -87	PCF86 -57	1.80	VR105 -50		
	ECH83		EY88 55	PCF200 1-20	QV04/78-00		ACY22 -18	
,	ECH84			PCF2011-00		VU111 1.00	ACY28 21	
_	ECL80		RY91 .50	PCF801 -49	QV06/20	VU120 1.00	AD140 -42	
60			EZ40 -52	PCF802 -54	8.50	VU133 1.00	AD149 -58	
70	ECL82	-40	BZ41 -52	PCF806 -58	R19 .75	W107 -75	AD161 -58	
00	ECL83	-74	BZ80 -32	PCH200	UABC80 -45	W729 1.20	AD162 -58	
70	ECL84	-65	EZ81 -32	1.00	UAF42 -70	X41 1.00	AF114 -80	OA70 -18
60	ECLA5	.70	GY501 -85	PCL82 40	UBC41 -50	X66 1.60	AF115 -80	OA73 18
BO	ECL86	-50	GZ32 ·60	PCL83 -49	UBC81 -55	Z759 5-85	AF117 -23	
75	EF22	1.00	GZ33 1.80	PCL84 -46	UBF80 -50	1	AF121 -35	OA81 11
85	EF40		GZ34 ·75	PCL86 -54	UBF89 -89	Transistors	AF124 -86	
45	EF41	-75	GZ37 1-10	PCL805 -60	UBL21 2-00	& Diodes	AF180 -56	OA90 14
50	EF42	-90	HABC80 -80		UC92 ·50	1N4744 ·16	AF186 64	
00	EF73	1.75	HL23DD-68	PEN25 1-00	UCC84 -90	2N404 ·21	BA115 -16	
	EF80	-29	HL41DD	PRN45 1-00	UCC85 -45	2N966 -61	BA116 21	OC44 -12
09	EF88	1.25	1.00	PEN45DD	UCF80 -80	2N1756 -58	BA129 -14	
09	EF85	-36	HL42DD	1.00	UCH21 2-00	2N2147 -99	BA130 -19	
70	EF86	-45	1.00	PEN46 -60	UCH42 -71	2N2297 -26	BA153 18	
L5	EF89	- 32	HN309 1.70	PEN453DD	UCH81 -45	2N2369 -16	BC107 -14	
	EF91		HVR2 1.00	2.00	UCL82 45	2N3053 38		
00	EF92	-60	HVR2A1-00	PENDD-		2N3121 2-90		
	EF97	-90		4020 1.00			BC109 14	
50	EF98	-90	KT41 1-00 KT66 8-00	PFL200 -70	UF41 ·70	2N3703 -23 2N3709 -23	BC113 -30	
30	EF183	-36		PL33 -50	UF42 ·80		BC115 -18	
10	EF184	-86	KT81 2.00	PL36 -60	UF80 -40	AA119 -18	BC116 -30	
30 I			KT88 6.75	PL81 -49	UF85 -50	AA120 ·18	BC118 -26	
10	EF804	1.75	KTW61		UF89 -45	AA129 -18	BCY10 -53	
5.5	EH90	-45	1.50		UL41 -70	AAZ13 -21	BCY12 - 58	
70	EL32	-60	KTW62	PL82 -87	UL84 ·48	AC107 -18	BCY33 -23	
75	EL34	-90	1.50	PL83 45	UM80 -60	AC113 -30	BCY34 -26	
ŏ	EL35	8.00	KTW68	PL84 .50	UY41 -50	AC126 -14:	BCY38 -26	
7	EL37	8.00	1.20	PL504/500	UY42 -50	AC127 -20	BF158 21	OC83 -23
	EL41	-57	MHLD6	90	UY85 -35	AC128 ·23	BF159 -30	OC84 -28
75	EL81	-65	-99	PL505 1.55	U10 1:00	AC132 -23	BF163 -23	OC123 -26
15	EL83	.70	MU14 1:15	PL508 1.00	U12/14 1-15	AC154 -80	BF173 -44	
10	EL84	-34	P61 -60	PL509 1.55	U19 4.00	AC156 -23	BF180 -35	
15	EL86	-60	PABC80 -45	PT4D 1.00	U25 ·71	AC157 -80	BF181 -47	
10	EL95	-67	PC86 -62	PY33/2 -50	U26 -60	AC165 -80	BF185 -47	
100	EL360	1.80	PC88 -62	PY80 -50	U33 1.75	AC166 -80	BFY50 -26	
14.	EL506		PC92 ·55	PY81 -40	U35 1.75	AC168 -44	BFY51 -22	
4	EM80		PC95 -70	PY82 40	U37 2.00	AC176 -64		
5	EM81		PC97 -89	PY83 44	U45 1.20			
10	EM83		PC900 -40	PY88 40	U81 -80	MATCHED T	HANSISTOR	SETS:-
io I	EM84	-45	PCC84 -89	PY500A1.20	U191 50	LPID (ACII3	. AC154. AC1	57. AA120) .61p
0	EM85		PCC85 47	PY800 -40	U251 1.00	per pack, 1/0	C81D & 2/0	C81, 50p.
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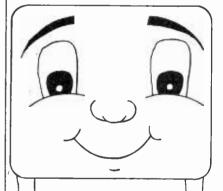
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This kit is ready to assemble and contains all ints at its ready to assemble and contains all genulue short-wave components, drilled chassis, valve, accessories and full instructions. Full range of other 8.W. kts. including the fannous model "K plus" (illustrated above), All orders despatched within 7 days. Send now for free descriptive catalogue of kits and components. Send stamped envelope for details.

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14 watts peak into 8 ohms. 2 x 8"approx chassis speakers and BSR C141 auto Personal shoppers only record player deck.

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RMS per channel output, full short-circuit and overheat protection. LIST 2495 PAIGE

Complete with PZ20 Power Supply

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FASY-TO-BUILD WITH ENCLOSURE

Specially designed by RT-VC for cost-conscious hi-fi enthusiasts, these kits incorporate two teaksimulate enclosures. two EMI 13"×8"

(approx).

(approx.) woofers, two tweeters and a pair of matching crossovers. Easily constructed, using a few basic tools. Supplied complete with an easy-tofollow circuit diagram, and crossover rms. 30 watts peak, each unit. 2550 Cabinet size 20"×11"×91/2"

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+ p & p £5 50

How about this for incredible bookshelf value from RT-VC! A pair of high efficiency units for only £7.50 – just what you need for low-power amplifiers. These infinite baffle enclosures come to you ready mitred and professionally finished. Each cabinet measures 12"×9"×5" (approx.) deep, and is in wood simulate (approx.) speakers for max. power handling of 7 watts. + p & p £1 70

50 PER SET 15-WATT KIT IN + p & p **CHASSIS FORM** 

When you are looking for a good speaker, why not build your own from this kit. It's the unit which we supply with the above enclosures. Size 13"×8" (approx.) woofer (EMI), tweeter, and matching crossover. Power handling capacity 15 watts rms 30 watts peak.

#### **DECCA 20 WATTS** STEREO SPEAKER

This matching loudspeaker system is hand-made, as only Decca know how, built to a specification, not down to a price.

The kit comprises of two 8' diameter approx, base drive unit, with heavy die cast chassis laminated cones Our price with rolled PVC surrounds, per stereo pair Two 3½" diameter approx. domed tweeters complete with crossover networks.



53000

## 20 x 20 WATT STEREO AMPLIFIER



Superh Viscount IV unit in teak-finished cabinet Silver fascia with aluminum rotary controls and pushbuttons, red mains indicator and stereo jack socket. Function switch for mic, magnetic and crystal pick-ups, tape, tuner. and auxiliary. Rear panel features two mains outlets, DIN speaker and input sockets, plus fuse, 20 + 20 watts rms, 40 + 40 watts peak

### HOW YOU CAN SAVE

SYSTEM 1B For only £80, you get the 20+20 watt Viscount IV amplifier, a pair of our 12-watt-rms Duo Type Ilb matched speakers; a BSR MP 60 type deck complete with magnetic cartidge, de luxe plinth 580°0 and cover.

p & p £6 50

SYSTEM 2 Comprising our 20+20 watt Viscount IV amplifier: a pair of our large Duo Type III matching speakers which handle 20 watts rms each, and a BSR MP 60 type deck with magnetic cartridge, £9200 de luxe plinth and cover. . 08 p 57 60 Carnage surcharge to Scotland System 1B - 2 50 System 2 ±5 00

SPEAKERS Two models- Duo IIb, teak veneer, 12 watts rms, 24 watts peak, 181/2" × 131/2" × 71/4" approx 234 PER PAIR - p.8 p.26 50

Duo III, 20 watts rms, 40 watts peak, 27" × 13" × 111/2" approx E48 PER PAIR p8p €7 50



TURNTABLE Popular BSR MP 60 type, complete with magnetic cartridge, diamond stylus, and de luxe plinth and cover

£2400



## 30 x 30 WAT

Specially designed by RT-VC for the experienced constructor, this kit comes complete in every detail. Same facilities as Viscount IV amplifier.

Chassis is ready punched; drilled and formed Cabinet is finished in teak veneer. Black fascia and easy-to-handle aluminium knobs. £2000 Output 30+30 watts rms, 60+60 peak. p & p £2 10



**BSR T145** 8-TRACK CARTRIDGE

PLAYER MECHANISM Requires some atten-

tion. Complete with built in pre-amp, A.C. 240V

p & p €1 50



#### 35-WATT DISCO AMP

Here's the mono unit you need to start off with. Gives you a good solid 35 watts rms, 70 watts peak output. Big features include two disc inputs, both for ceramic cartridges, tape input and microphone input. Level mixing controls fitted with integral push-pull switches Independent bass and treble \$2750 controls and master volume.



#### PORTABLE DISCO CONSOLE with built-in pre-amplifiers

Here's the big-value portable disco console from RT-VC! It features a pair of BSR MP 60 type auto-return, single-play professional series record decks. Plus all the controls and features you need to give fabulous disco performances. Simply £5500 connects into your existing + p&p£6 50 slave or external amplifier.



#### 70 & 100 WATT DISCO AMPS

Brilliantly styled for easy disco performance! Sloping fascia, so that you can use the controls without fuss or bother. Brushed aluminium fascia and rotary controls. Five smooth-acting vertically mounted slide controls - master volume, tape level, mic level, deck level, PLUS INTER-DECK FADER for perfect graduated change from record deck No. 1 to No. 2, or vice versa. Pre-fade level control

(PFL) lets YOU hear next disc before fading it in. VU meter monitors output level 70 watts rms, 140 watts peak output. All the big features as on the 70-watt disco amplifier, but with a massive 100 watts rms. 200 watts peak output power.

70 WATT £4900 - p&p£3 00 100 WATT

£6500 - p & p £4

### TOURIST IV PUSH BUTTON R RADIO KIT



#### MOTOR TOP 10 AWARD

Complete with speaker, baffle and fixing strip. The Tourist IV for the experienced constructor only. The Tourist IV has five push buttons, four medium band and one for long wave band. The tuning scale is illuminated and attractive small aluminium control knobs are used for manual tuning and volume control.

The modern style fascia has been designed to blend with most car interiors and the finished radio will stol into a standard carradio aperture Size approx 7" × 2" × 43". Power Supply Nominal 12 votts positive or negative earth (aftered internally) Power Ouptut 4 watts into 4 ohms

, 10°



### STEREO CASSETTE DECK KIT

Again, this kit is specially designed for the experienced constructor-for mounting into his own cabinet. Features include solenoid-assisted AUTO-STDP, 3-digit counter, record/replay

PC board, mains transformer and input \$3250 and output controls. AC BIAS AND ERASE. p & p £1.50

**DELUXE ACCESSORY KIT** Comprises of a matched pair of dynamic mics, and two replacement slider level controls. This item post POST FREE when purchased with Cassette Deck kit

£**3**95 p&p£1.00

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card number

## AMBIT international (dept 85)

The Dynamic Twosome: Signalmaster/Audiomaster After long and thorough deliberation, we are proud to announce a new unit from Larsholt - the Audiomaster. As ever, the instructions are designed to lead the unwary-and the inexperienced through point-to-point steps that culminate in a professionally styled and finished amplifier to complement the Signalmaster FM tuner. Price £79.00



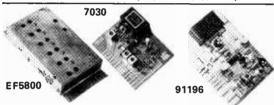
Power: 25+25W RMS THD: Less than 0.3% Dynamic range: an exceptional 80dB (Signalmaster shown on top of the Audiomaster)

The Signalmaster Mk.8 is equally simple to assemble, and results reflect the superb Scandinavian styling and careful electronic engineering. £85.00.



International Mk.2: A choice of tuners for the more experinced constructors.

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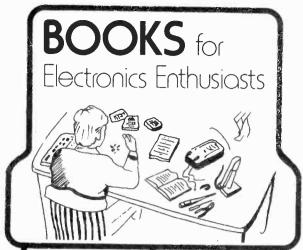
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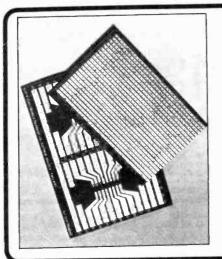
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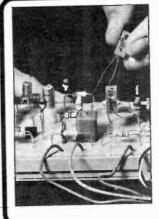
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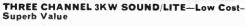
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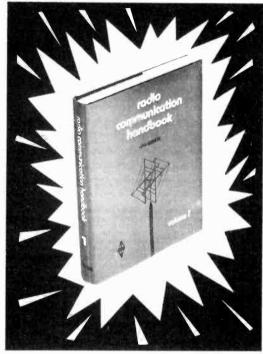
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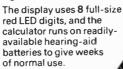
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## JDA HANI

It may be supposed that the great majority of our readers are reasonably sound in wind and limb and able to move around, more or less freely, under their own steam. They can look at a project in the magazine, decide to make it and then jump into their car, or take a bus, and go and get the parts that they need. If they are very lucky and don't live in London or one of the other big cities then they can pop down to the local postbox and send

off their order to one of our advertisers.

Having just received the latest copy of Radial, the magazine of the Radio Amateur Invalid and Bedfast Club, I am reminded that there are those among us who are not quite so fortunate. Some who are very enthusiastic licensed radio amateurs or listeners but who are handicapped in one way or another and must rely upon others to set up and maintain their gear and aerials. But can you imagine the joy that these people derive from being able to communicate with or listen to other amateurs from all over the world?

The Club is very active and prospering under the Presidency of Bill Browning G2AOX and his committee but it must be remembered that it is the ordinary healthy member who is so important and on whom so much depends. One such recent recruit was Ken Alford G2DX (what a fabulous call!) who at 83

feels he ought to be helping the youngsters, no doubt!

How many of our readers have bits of equipment or, in particular, short wave receivers, knocking about in the attic or garden shed that could be put into good working order and handed over to the RAIBC without a great deal of trouble? Many, for sure. So even if you don't want to join the Club you can still help in a very practical way those who are generally unable to help themselves.

If you do help you will certainly feel a little glow of satisfaction after having done your bit and may even be tempted to carry on the good work. It might even mean a couple of hours less spent working the DX from your own rig but it will be worth while. After all, conditions aren't so good all the time! If you feel that you want to become involved write in the first instance to RAIBC Secretary Harry Boutle G2CLP of 14 Queen's Drive, Bedford and

he'll put you on to the right track. Now, where's that loft ladder? If you'd like to hear the RAIBC boys on the air try 3750kHz SSB at 1000 on Tuesdays or 1400 on Wednesdays when the Net controller is G3WJT or perhaps the Cheshire Homes Net around 3650 to 3700kHz SSB at 1330 on Thursdays when the MC is

G3OPY.

Member Tony G4BPX summed it up very well, in an article in Radial, when he said 'In thinking with and for less able friends for over twenty years I have become less handicapped myself' Think about it!

Eric Dowdeswell—Assistant Editor

This issue contains the first of the free PW Info-Cards, on Resistors and Capacitors, see page 938. The second Info-Card will be given away with the May issue of PW and will deal with Semiconductors.

DON'T MISS IT!

#### **Polling time**

HE International Short-Wave Club is now holding the tenth in its series of official polls which have been held every three years since 1950 to determine, in the listeners opinion the most popular short-wave Broadcasting stations. The poil began on November 1st, 1976 and will continue until February 28th, 1977. In order to enter, write on a postcard your five favourite Short-Wave stations, in order of preference, and add in not more than 30 words, the reasons for your first choice.

Results will be published in the April issue of the ISWC and announced by many stations. Further information from Mr G. C. Gibbs, 118 Bournemouth Park Road, Southend-on-Sea, Essex, SS2 5LS. Please enclose an SAE with all correspondence.

#### New licence explained

ROM January 1, 1977, the Home Office has been issuing four new types of radio amateur licence which have replaced all existing amateur licences when they become due for renewal.

The new Amateur Licence A (full facilities) and B (having certain limitations) will include all those facilities available to holders of the existing Amateur (Sound) Licences A or B, and will also include operation from a vehicle, or vessel other than at sea, or as a pedestrian. Fascimile, amateur television, slow scan television, data on amateur bands 144-146 MHz and above, as well as double sideband suppressed carrier operation will also be included. **Emergency** County Planning Officers will also be added to the categories able to call upon Raynet (emergency communications).

The purposes of the new-style licences are:

(1) to give more flexibility to

## NEWS...

## NEWS...

## NEWS...

British radio amateurs, so that they may pursue their hobby without having to make special application for several of the above facilities at present needing separate licences or authorities.

(2) to enable the Home Office to deal with the increase in applications for amateur licences and regulatory work over the next few years with the minimum delay.

Existing facilities authorised by a total of 20 licences and special authorities will thus be contained (with a few exceptions) in the new Amateur Licence A or B; and for Aliens combined fixed and mobile facilities in the Amateur Licence C or D. The new fee for all UK amateur licences will be £5 50.

#### Catkin, Red E, Trigatrons . . . !

T a recent meeting of the University College of North Wales ARS John Lawrence, GW3JGA, gave a talk on the "Development of the Radio Valve" and showed part of his collection of historic radio valves. On display were over 200 valves covering the 60 year period from 1915 to 1975.

The pre-war range included an "R" triode, AOR triodes, early indirectly-heated valves, screen grid types, "Catkin", Mullard "Red E" and Mazda octal versions. The wartime range included klystrons, trigatrons, magnetrons (including a MAG1) and a range of unusual disc-seal triodes which were used in early radio-location equipment. The wartime and post-war types included many transmitting valves, one of the latest vapour cooled triodes on show being rated at 60kW CW output.

Smaller types such as acorns, deaf-aid valves and Nuvistors were also included. In addition to valves a number of opto-electrical devices, photo-cells, camera and convertor tubes were on display. To give a cheery atmosphere to the meeting, two large VT31 tetrodes were set up with their directly-heated filaments glowing brightly (filament rating 10V 10A).

### Navy days

ACTIVITY period from HMS

Belfast, in the Pool of London, 8th to 17th April. Call
sign will be GB3RN.

Mobile Rally at HMS Mercury, nr. Petersfield, Hants. Sunday, 19th June.

Navy Days at Portsmouth and



Practical Wireless, March 1977

Plymouth dockyards, 27th, 28th and 29th August. GB3RN and GB3GUZ.

Jamboree-on-the-Air Camp and Radio Teach-In at HMS *Mercury*, 15th and 16th October.

Annual General Meeting at HMS *Mercury*, nr. Petersfield, Hants, 22nd October.

For further details or membership details of the RNARS write to M. J. Matthews, G3JFF, Hon. Secretary, HMS *Mercury*, East Meon, Petersfield, Hants.

#### **Spring show**

A SHORT note here for those readers who attend, or would like to attend the All-Electronics Show which is held annually. This year's show promises to be as successful as those in the past, and will be held at Grosvenor House, Park Lane, London W1 on the 19th-21st April. Watch this page in the coming months for more news on this exhibition.

#### Teach-in

THE Department of Electrical Engineering Science at Essex University will be holding its annual Electronics Summer School for teachers during the week July 11-15, 1977. Two courses will be run simultaneously. The Linear Circuit Design course is concerned with the use of transistors and operational amplifiers in analogue applications and the basic circuits of a HiFi amplifier will be investigated in detail. The Digital Circuit Design course concentrates on the use of the transistor as a switch and develops design using integrated logic circuits.

Further information from: R J. Mack, Department of Electrical Engineering Science, University of Essex, Wivenhoe Park, Colchester CO4 3SQ, telephone Colchester (0206) 44144, extension 2408 or 2299.

# Circuits for AUDIO AMPLIFIERS

## PART 1

F.G.RAYER G3OGR

THE circuits in this series are intended to give practical essential information for the construction of pre-amplifiers, mixers, tone-controls, low-level and Class A output stages as well as various types of push-pull output stages and IC audio amplifiers. Some circuits can be added to existing equipment, or require few components and can be constructed as separate units. Others are more ambitious.

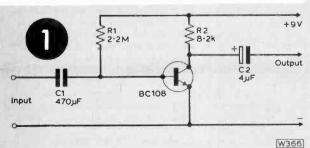
The simpler circuits can be assembled on tag-board but for the larger circuits Veroboard can be used, or printed circuit boards can be prepared. Copperclad board and the other necessary materials for this purpose are readily obtainable. Numerous cases and cabinets suitable for housing the finished equipment are available and they can include space to accommodate the battery or AC power pack, as appropriate. Input and output sockets and controls can be mounted on the case panel, as well as brackets to carry the circuit board.

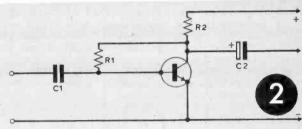
The circuits here employ both negative and positive "earth" or chassis line. This will prove to be convenient when new equipment has to suit existing units. Some circuits are readily adapted for either negative or positive earth though, in general, the positive earth is more likely to be found with older amplifiers employing PNP transistors.

#### LOW-LEVEL "CLASS A" STAGE

A Class A stage is one which is required to amplify the whole audio cycle. The output from the stage should be as faithful a copy of the input as possible, unless some deliberate frequency correction has been made. A pre-amplifier or early audio stage should introduce as little noise as possible, as this would be amplified by later stages. Typical Class A low-level stages include pre-amplifiers and the early stages of an amplifier.

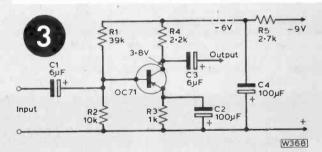
Fig. 1 is a simple low-level Class A stage. Resistor R1 furnishes base bias and the NPN transistor is





W367

used as a grounded emitter amplifier. This particular circuit can be used by itself or in simple equipment, but lacks any form of stabilisation. As a result, collector current and operating conditions could vary considerably due to spreads in transistor characteristics, changes in temperature or variations in voltages. It is thus necessary with this circuit to limit base and collector currents to low levels. This circuit can be modified by taking R1 to the collector, Fig. 2. This provides a useful degree of stabilisation since base current through R1 depends on collector voltage, as well as on the value of R1. Component values may be as in Fig. 1.



The circuit of Fig. 3, which has a positive earth

line, is stabilised. Resistors R1 and R2 form a potential divider setting the base operating point. R3 is the emitter bias resistor and R4 the collector load. With the PNP transistor, the base will be slightly negative, relative to the emitter. Should transistor spread, supply voltage or a rise in temperature cause collector current to be unusually high, the voltage drop across R3 rises due to the increased emitter current. The emitter moves negative, so that

the emitter than before, thus reducing collector current.

Should conditions result in an unusually low

the base is relatively less negative with respect to

collector current, the reverse happens. The voltage

drop across R3 is reduced. The emitter thus moves positive, so that the base is relatively more positive and this tends to maintain collector current. In this way stable, satisfactory working conditions are maintained.

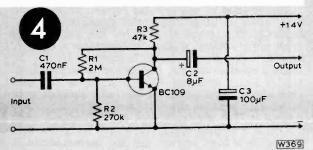


Fig. 4 is another low-level amplifier, using a lownoise NPN transistor and it may be used to boost low-level inputs. The 14V supply line is not essential and R3 may be reduced in value with lower voltages, good results being obtained with 6V or less if necessary.

#### TONE CONTROLS

Tone controls may curtail or emphasise treble or bass, according to the circuits used, and are employed to obtain more satisfactory tonal balance or to compensate for recording characteristics, speaker deficiencies or other factors.

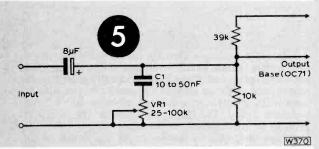
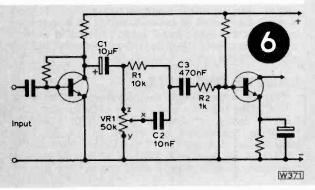
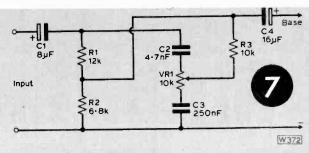


Fig. 5 is a simple top-cut control, treble response being curtailed as the value of VRI is reduced. The values shown are not critical. Such tone controls should not be added to a stage over which negative feedback takes place but to an early stage outside the feedback loop.

The reactance of a capacitor falls as frequency rises. In Fig. 5 Cl has a lower reactance to high frequencies, so that their amplitude is reduced at the transistor base. This change of capacitor reactance

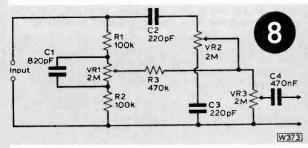




with frequency is generally used in tone control circuits. Cutting treble and turning up the amplifier gain effectively increases the bass frequencies.

Fig. 6 is a circuit in which the potentiometer VR1 allows top frequencies to be cut when slider X is moved towards Y (similar to Fig. 5). However, when X is moved towards Z, C2 is increasingly effective across R1, to lift treble response. These and similar circuits can often be placed in existing equipment. Base, emitter and collector resistors are as for the stage without tone control.

Fig. 7 is another tone control for an early stage. R1 and R2 form an audio potential divider for C4 and the base of the following stage. VR1 provides best bass reproduction when adjusted towards C3 and best treble response when towards C2. Fig. 8 also controls bass and treble but has individual potentiometers. It can be fitted before the first stage if required. VR1 is the bass control and VR2 is for treble. VR3 is the volume control. This is a purely



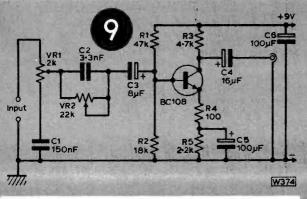
passive circuit which may be assembled in a metal screening box and requires no battery.

Circuits such as these can often be added to existing amplifiers without difficulty as they do not upset the DC working conditions. Some extra gain may be needed to compensate for losses in the tone control circuit when this does not incorporate a transistor. All values can be considerably modified, though there is no point in providing extremes of bass or treble which will never be wanted.

#### PRE-AMP WITH TONE CONTROL

Fig. 9 is a circuit of a pre-amplifier which incorporates two variable tone controls, VR1 is for treble cut and VR2 for bass cut. A volume control may be incorporated between VR1 and the input circuit, if required. Such a stage is useful when the main amplifier lacks sufficient gain for some particular purpose. This circuit also allows tone controls to be provided as well, with no changes to the main amplifier. If a pre-amplifier is operated from its own battery, no difficulties of hum or coupling through the power supply can arise.

When operation is required from the same supply as the main amplifier, a decoupling resistor is added



in the positive supply line, between C6 and the supply circuit. As most amplifiers will be operated from the mains these components (resistor and C6) also help to assure a smooth, hum-free supply for this stage.

A pre-amplifier can have circuits designed for various inputs such as for radio, or for moving-coil or crystal microphones. These may be selected by switching.

#### INPUT IMPEDANCE

Tuners, crystal pick-ups and other inputs are best operated into loads of a particular impedance. This is not too critical for general purposes but becomes important for improved fidelity. A suitable impedance is frequently obtained with a resistor network. It causes a loss in signal strength, but this is easily restored by a pre-amplifier, if necessary.

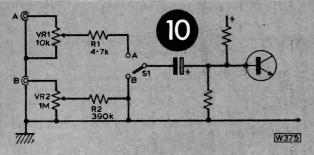


Fig. 10 is a circuit having two inputs, "A" is of relatively low impedance for a radio tuner, preamplifier and other inputs of around  $10k\Omega$  impedance. Input "B" is more suitable for a crystal pick-up and other high impedance inputs, where a low impedance would reduce frequency response. Individual volume controls VR1 and VR2 can be set to suit the signal levels provided.

#### INPUT LEADS

The microphone, pick-up or other input leads to the pre-amplifier and amplifier must be screened. So must the inter-connecting audio circuit leads from pre-amplifier, tuner or other units to the main amplifier. The purpose of such screening is to avoid pick-up of hum or other currents induced from adjacent circuits, which, if introduced into the input circuit would be amplified and accompany the signal or cause instability.

The screened leads are fitted with jack plugs or other audio-type connectors to match the equipment.

The outer conductors of the leads form the earthed or chassis return, but should not be used to carry power between units.

#### "CLASS A" OUTPUT

The addition of one or more further stages to the first, low-level stage raises the signal level, eventually to that required to drive the loudspeaker. With a dry battery, a single transistor as a Class A amplifier can provide ample headphone volume, or modest speaker volume, mainly for personal listening. A single transistor Class A amplifier can give very much higher power in car radio equipment or in mains-operated circuits where current is not limited to that available from dry batteries. Powerful Class A circuits require an output transistor able to handle a high current and the current drain is greater than with an equivalent power Class B output stage (described later). For this reason large Class A stages are not used in battery operated receivers or amplifiers.

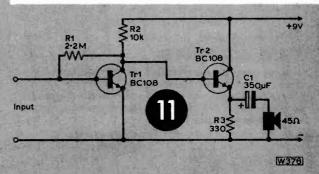


Fig. 11 is a simple directly coupled 2-stage amplifier with a small Class A output stage. It provides excellent gain and gives a surprisingly good volume when used with an efficient speaker. It can also be used with phones or a midget speaker for a small personal radio. The input is suitable for a crystal pick-up or microphone. With other inputs, an isolating capacitor will be necessary in the input circuit so that bias conditions (via R1) for Tr1 are not upset.

Tr2 is an emitter follower, with grounded collector. The collector is earthed as far as audio signals are concerned, as the battery supply is considered to have zero impedance. DC operating conditions for Tr2 are not upset by the speaker resistance as C1 isolates the speaker, but results are best if the speaker is of moderately high impedance.

When a Class A stage is operating properly, a milliampmeter placed in the supply circuit should show no significant change of current with varying signals. But in small amplifiers like Fig. 11 there is some fluctuation of current, except at low volume levels. On the other hand, large changes in current are expected with Class B push-pull circuits and are quite in order.

#### LARGER OUTPUTS

Class A output stages running from accumulators, about 12 to 14V, are popular for vehicles and often have a rating of about 2 to 5W with a current drain in the region of 400mA to 1A. A large power Class A stage for 12 to 14V is shown in Fig. 12. It can be used with PNP or NPN transistors of the OC26 or BD121 type with appropriate polarity continued on page 927

# SPECIAL PRODUCT REPORT

## HEATHKIT



any thousands of radio amateurs were introduced to the thrills of operating with low power CW when Heathkit thought up the earlier HW7 transceiver. However that did not have the 80 metre band available, a sad omission in these days of low sunspot activity. There was no gain control on the RF side and there was separate dial calibration for each of the three bands. On the credit side, it was possible to plug in a crystal for fixed frequency operation, a facility lacking on the HW8.

But what is 'low power'? In the USA, 100W or so is spoken of as low power, which it probably is when one considers their maximum licensed power. I prefer to go along with our ideas over here, which is something below 10W input, although there are many enthusiasts who use, and get good results from, a few hundred milliwatts.

#### DESCRIPTION

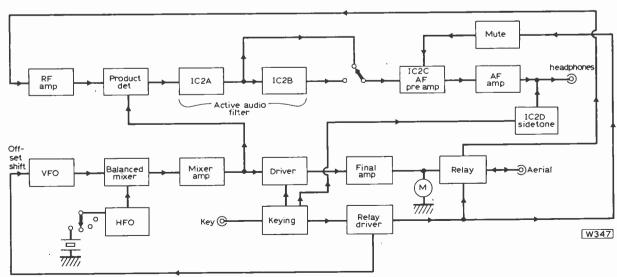
The HW8 transceiver is designed to work from a nominal 12VDC supply, in practice from a car battery if in the field or from a mains unit if at home or in a hotel on vacation. In

either case the actual input voltage will be nearer to 13·4V. On transmit the HW8 provides an input of around 3W on CW on the bottom 250kHz portion of the 3·5, 7, 14 and 21MHz bands. The output impedance is fixed at 50 $\Omega$  with an output loading control on the front panel, coupled with a relative output meter. The receiver uses the direct conversion technique, as did the HW7, and as well as having an RF stage it has a switchable two-stage active filter giving audio bandwidths of 375 and 750Hz (6dB) at 750Hz. In use this means that SSB is just about uncopyable which is rather a pity, although only CW can be transmitted. The two modes are far from incompatible.

#### THE CIRCUIT

On both receive and transmit the VFO output (8.645 to 8.895MHz) is mixed with the output of an RF oscillator which provides four crystal controlled frequencies, 12.395, 15.895,

This block diagram shows how several of the stages of the HW8 are used on both transmit and receive.



### abridged specification

TRANSMITTER

RECEIVER

DC Input 80m 3.5W

40m 3·0W S 20m 3·0W pr 15m 2·5W S

Direct conversion, CW only Sensitivity 1,µV with 0·2,µV providing readable signal Selectivity Wide:—750Hz @ 6dB down. Narrow:—375Hz @ 6dB down. Passband Freq.

Impedance 50Ω unbalanced

Output

750Hz Output Imp. 1kΩ

Transmit Offset 750Hz lower

GENERAL

Frequency Range MHz 3·5 to 3·750, 7·0 to 7·250

14 · 0 to 14 · 250, 21 · 0 to 21 · 250

Frequency Stability

Less than 150Hz/hour after

1 hour

**Power Required** 

13.4VDC nominal 90mA/ receive 430mA/transmit 9½ x 4½ x 8½in. deep overall.

Dimensions

(235 x 108 x 216mm)

Weight

4lbs. (1.8kg)

22.895 and 29.895MHz. The difference frequency is selected for each band by tuned circuits, thus giving 3.5 to 3.750, 7.0 to 7.250, 14.0 to 14.250 and 21.0 to 21.250MHz. The dial calibration of 0 to 250 is thus constant for each band. These frequencies are also used as the heterodyne frequencies for the direct conversion receiver.

This is a good point at which to say that there are 20 tuned circuits plus the four crystals all of which are diode switched! The band switch thus operates at DC rather than at RF, except for the PA tuned circuits. It is not sumprising then that the main PCB carries some 300 or so components!

There are 32 coils, 37 diodes, 92 resistors, 120 capacitors and 16 transistors/IC's. Whether all this sophistication is justified to provide a couple of watts of CW on four bands is a matter of opinion!

On transmit the output of the mixer is amplified and fed to a tuned driver stage into the 2N4427 power amplifier. This stage is protected against excessive SWR conditions or accidental no-load dangers. Keying takes place in the driver stage, also activating the transmit/receive relay, the keying sidetone and the receiver muting circuits. There is an adjustable drop-out time control on the T/R relay.

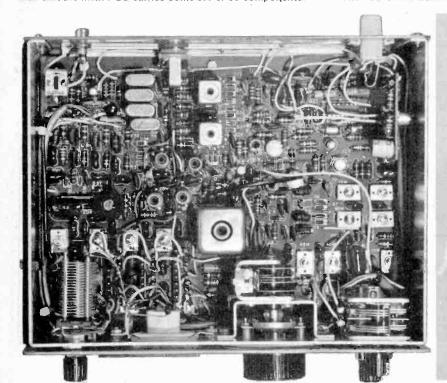
The receiver has an RF stage with the simplest of gain controls, which many amateurs, who complain of cross-modulation on their sets, would do well to copy. Namely, a potentiometer across the input terminals! An MC1496G IC forms the balanced product detector, also fed with a signal from the mixer stage, as previously mentioned. The active audio filter stages have also been covered earlier. Because of the 750Hz peaking effect of the audio filters the carrier has to be offset 750Hz on transmit. This is done automatically when the key is down. In practice this means tuning down from the HF end of a band and tuning a signal right on the nose' of the filter giving a 750Hz beat note. The transmitted signal will then be spot on the frequency of the incoming signal.

Two further stages of audio amplification feed a headphone socket at high impedance, nominally  $1k\Omega$ , but in practice a stereo low impedance headset performed quite well But an old pair of Browns diaphragm type of  $4k\Omega$  really got the signals pumping out when peaked on the filter frequency!

#### CONSTRUCTION

As with any kit the components supplied must be checked against the parts list. Every part, down to the nuts and bolts, is illustrated in the Heathkit manual which is an education in itself and even before construction begins a good knowledge of the HW8 is obtained. Personally, I always read the manual right through anyway, before ever lifting the soldering iron! It is time well spent, often explaining points that might otherwise prove mystifying during construction.

The bulk of the work is on the main PCB which carries no



Top view of the finished transceiver. All the cabing round the inside of the cabinet is fitted to the PCB before the cabinet is assembled round the board. The TIR relay is in the top left hand corner close to the aerial socket, with the four HF crystals further to the right. The large screened can in the centre is the VFO unit with the VFO tuning immediately in front on the panel. While the PCB may appear rather cluttered up, the wiring in of the components presented no difficulty at all. There are no components under the PCB which is in the bottom of the cabinet.

less than 308 components! Actually the lob of wiring so many resistors and diodes etc. gets boring so give it a break now and again to avoid making errors. The PCB and the associated cabling took nine hours to complete. It is an excellent board on which to work and need not be any cause for apprehension for potential constructors. The PCB was carefully checked for solder 'bridges' with the aid of a watchmaker's eyeglass and all was well.

The simple cabinet, front and rear panels are fitted round the PCB and the wiring completed to the various controls on the front and sockets on the rear panel. The aerial socket is a 'phono' type easily changed to Belling-Lee type or to suit the rest of the station's RF cabling. After completing the PCB a further 5½ hours work saw the transceiver completed. A further check on the PCB underneath and the general

wiring was made before tests began.

#### TESTING

Although the manual did not mention it, it was thought advisable to check the resistance across the 12V input terminals just in case there was a short there! The multimeter red prod (neg polarity) was connected to negative input, giving a reading of 550 $\Omega$  with the transceiver switched on and on receive. The power unit used for testing was the Heathkit HWA-7-1 giving approx. 13 2V. The initial tests laid down in the manual all proved satisfactory. The relay worked on keying, the sidetone was present and the receiver controls worked but little RF output was present at this stage.

The test equipment required is a receiver, preferably of the communication type, capable of providing accurate checks at 7 and 7 250kHz for checking the VFO, and a high input impedance voltmeter preferably with an RF probe. The allgnment is quite clearly described and it is a matter of logical progression which can be completed easily in haif

an hour.

If one runs into trouble there is plenty of advice in the manual including PCB layout drawings with the voltages to be expected at various points. With a stabilised input voltage of 13VDC the current on receive was 93mA and on transmit 450mA, remaining fairly constant over the four bands. The RF output was measured at the dummy aerial load with the dial set at mid position and the loading control peaked and was found to be 2W on 80m and 40m, 1.8W on 20m and 1 6W on 15m.

Since the accepted segments of the bands on CW are a lot less than the 250kHz coverage available on the HW8 it would seem feasible to align the circuits at, say, 50kHz from the band edge and improve the general performance on both transmit and receive. In practice with the recommended alignment the RF output is remarkably constant over the

whole 250kHz.

#### GENERAL COMMENTS

The transmitted signal was monitored externally on all four bands and found to be clean with excellent keying characteristics. In fact it was possible to get the relay to follow keying up to about 25wpm although some may prefer to increase the relay drop-out time to a second or so. Plenty of DX has been heard on all four bands and the two position audio filter can really sort out the weak ones among the QRM. The RF gain control was most effective especially with some of the powerful signals on 80m.

#### AERIALS

Nothing spectacular has been worked on any band yet due to lack of sultable aerials. It is a problem to get a four band aerial which provides a low impedance feed on all bands, without using a tuning unit. Anything introduced between the aerial and the set is going to absorb some of the precious power. A trapped dipole is the quick answer but it is a compromise and rather 'lossy'. It is hoped to erect a four band ground plane with a coax feeder when a further report on results obtained will be published. AED

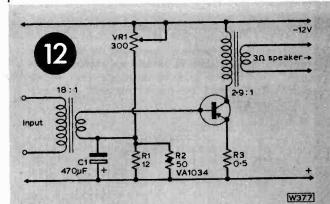
Heathkit k/HW8 Price £108 19 inc. VAT and post and packing from Heath (Gloucester) Ltd., Gloucester

GL2 6EE.

#### CIRCUITS FOR AUDIO AMPS .- contd. from page 924

of supply, that is positive earth for PNP and negative earth for NPN devices. It can be driven by an AC128 (PNP), BC108 (NPN) or similar stage. The transformer ratios are typical and need not be exactly as shown.

Collector current is set by VR1 and this can be the lowest which gives satisfactory performance. Begin with VR1 at maximum value, corresponding to a low collector current. Actual collector current depends on the transistor and performance and should be around 350mA to 900mA. As the DC operating conditions of this stage are isolated from other stages it is useful for experimental circuits and can actually be employed with about 9V to 24V. With the higher voltages, check carefully that the transistor is not over-run.



The load impedance can be around  $25\Omega$  but it is not too critical, depending somewhat on the supply voltage. Note that with a Class A stage a loudspeaker can only be connected directly to the collector circuit when it is able to carry the required collector current. This is why a transformer is necessary in Fig. 12.

Thermistor R2 is for stabilisation, as described later, whilst R3 also assists in this direction and can be about  $1\Omega$  if the volume is sufficient. Resistor R2 may be omitted for low voltages, or providing a check is kept that the current is not too high.

TO BE CONTINUED

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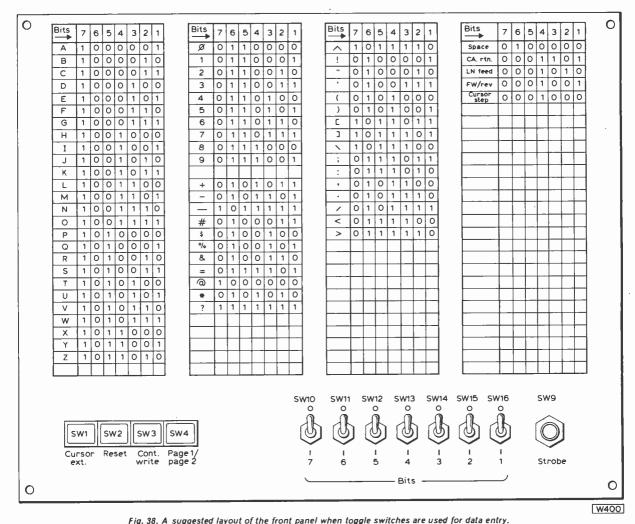
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## PART 8

## M.J.HUGHES M.A., C.Eng. MIERE

#### NON-KEYBOARD KEYING

Early on in the series we promised an alternative means of entering data to obviate an expensive keyboard. This is done with seven single pole toggle switches and a push-to-make button with an "antibounce" circuit. The code for the character or control function is set up on the switches in binary form and the button depressed to produce a strobe. If this method is used it is suggested that a table of the codes be prepared to tally with the switch positions. The suggested front panel layout and circuit are shown in Figs. 38 and 39. These codes might come in use-



nen toggie switches are used for data entry.

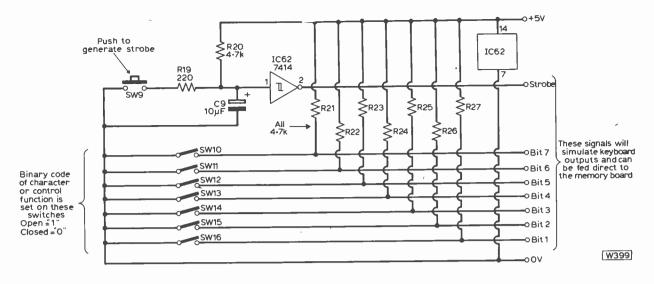


Fig. 39. The circuit diagram of the switching required to set data manually. A closed switch gives a logic "0",
an open switch a logic "1".

ful for anyone carrying out the trouble shooting procedure mentioned earlier.

Although this method of entering data will be very tedious for normal use it is ideal for setting up permanent or semi-permanent displays.

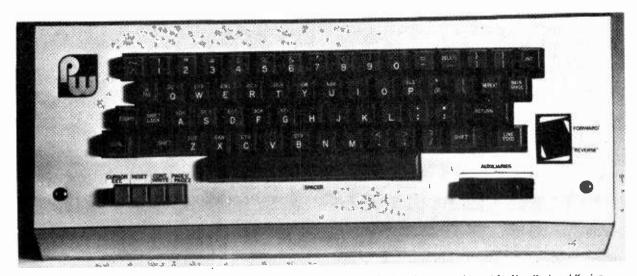
#### PAGE SEQUENCER

A page is selected by applying a logic level "1" or "0" to the most significant address bit of the random access memory. For page 1 this bit is held at logic "1" by R9 on the memory board. To change to page 2 the junction of R9 and the address input is shorted to ground (thus applying a logic level "0"). This is carried out by closing switch SW4 (page 1/2 select). This function could be carried out electronically by an oscillator and would give automatic page sequencing.

A variable-speed page-sequencing circuit can be added to the basic video-writer without any modifications to the original layout. The requirements are a small extra circuit board and a number of flying leads connecting it to the existing pinned "leadouts" on the main boards.

The circuit for such a sequencer is shown in Fig. 40 where IC62, a 555 operating in astable mode, is the oscillator and the frequency is set by VR1 and one of the three capacitors C9-C11. The range is set by selecting one of these capacitors via the auxiliary bank of switches (already embodied in the prototype).

VR1 gives fine control within the range selected and, in the prototype, is a preset but could be front panel mounted if desired. IC62 produces negative going pulses at the page sequencing rate and these are fed to the first part of IC63 which is a D-type



A photograph of the original keyboard. Note that the auxiliary switches are retained on the proposed layout for Non-Keyboard Keying.

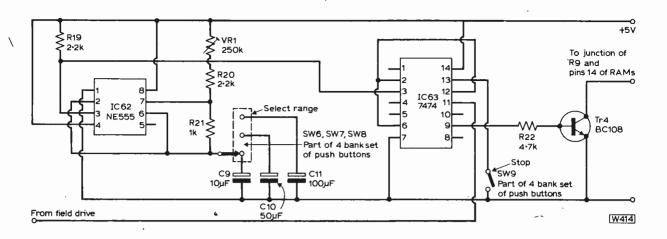
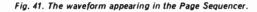
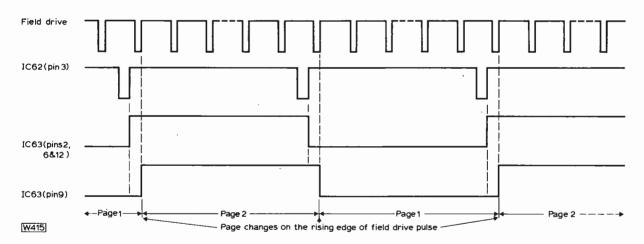


Fig. 40. The circuit diagram for the automatic page sequencer. SW4 must be open (i.e. switched to page 1) for this circuit to operate.

C9, 10 and 11 can be selected for any range of speed required.





flip-flop giving unity mark space output waveform.

It is desirable for the page to change during the field sync period. This prevents the displayed information changing while the scan is in the middle of a raster.

To do this it is necessary to synchronise the output at pins 2 and 6 of IC63 with the field drive pulse (obtained from the Sync Generator Board). This is done by aplying the output of the first D type flipflop in IC63 to the "D" input of its twin and using the Field Drive Pulse as a clock signal. The Clear input of this second half is connected to the fourth push-button of the auxiliary bank which serves as a "disable" to the page sequencer, (i.e. switches it off), and holds the output on pin 9 at logic level "0". The associated waveforms are shown in Fig. 41.

This signal has to be applied to the junction of R9 and the paralleled connections to pins 14 of the random access memories. This is most easily accomplished by the open collector transistor (TR4) which will pull the junction to ground when the output from the page sequencer is at logic level "1". Pro-

vided the manual page select switch SW4 is in an open circuit condition, TR4, in conjunction with R9, will allow the address to alternate from "1" to "0", according to the instruction from the sequencer unit.

If the disable button SW5 is pressed there is no base drive into TR4 and SW4 is free to give manual control of page selection. Note that for automatic page sequencing to occur, SW4 must be in an open circuit state otherwise the address will be permanently grounded (displaying "Page 2").

Using the values of capacitors and variable resistor in the parts list the duration of the page display can be varied from approximately 20 seconds a page to a fraction of a second a page.

High speed page strobing can produce interesting effects for advertising but should be used with discretion due to the hypnotic effects that can be induced. It is possible to strobe at the "Bio" Beta Frequency and under certain circumstances this can produce unpleasant side effects in a viewer.

The wiring pattern of the printed circuit board is shown as Fig. 42 and the component layout as Fig.



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Fig. 42. The etched wiring pattern of the Page Sequencer board drawn full size.

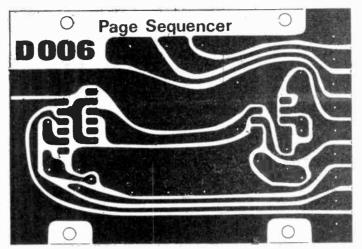
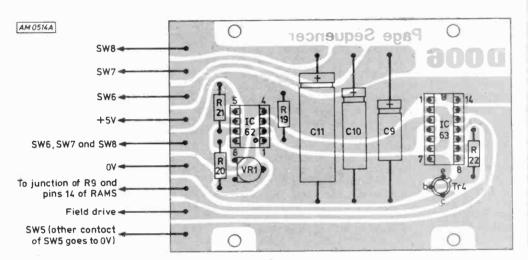


Fig. 43. The component overlay information. Observe the +ve marking when using polarised capacitors. VR1 is shown as a miniature preset but could be externally mounted.



#### **★** components list

#### NON-KEYBOARD KEYING 220Ω 10% ¥W R19 4·7kΩ 10% ½W 4·7kΩ 10% ½W **R20** R21 **R22** 4.7kΩ 10% 1W 4·7kΩ 10% ‡W 4·7kΩ 10% ‡W **R23 R24** 4.7kΩ 10% 1W **R25** 4·7kΩ 10% ‡W **R26** 4.7kΩ 10% 1W **R27** 10µF 12V 10% 1W C9 SW9 Push-to-make button switch SW10-SW16 Single pole toggle switches IC62 7414 Small piece of Veroboard.

43. The +5V and 0V supply lines are taken off the respective pins of the memory board; the output from the page sequencer goes to the Page 1/Page 2 input of the memory board; the Field Drive is taken from its associated pin on the Sync Generator Board and the page sequencer disable and range select inputs are taken to the respective contacts on SW5, 6, 7 and 8.

#### ★ components list

#### PAGE SEQUENCER 2·2kΩ, 10%, ‡W R19 2·2kΩ, 10%, ‡W R20 1kΩ, 10%, ½W R21 **R22** 4.7kΩ, 10%, 1W VR1 250kΩ (sub-min horizontal pre-set) 10μF (10V) 50μF (10V) C9 C10 100µF (10V) C11 TR4 **BC 108 NE 555** IC6 IC6 7474 One off 8 pin DIL socket One off 14 pin DIL socket SW5-SW8 Interlocking bank of 4 push switches Printed Circuit Board **Printed Circuit Board Pins**

This extra board can be mounted inside the Videowriter in any convenient position.

These two modifications complete the original article. Other add-on and interface units are under consideration and, subject to there being sufficient demand, will be published at a later date.



#### Thermionic passion

Being a long-standing reader, and not too old to remember valves, I really must protest about the dismissal of the Good Old Valve. December issue letters raised my passion. I rarely write protest letters but Mr. Harcourt must surely remember the Coherer, this being one of the earliest forms of detector and non-semiconductor.

Mr. Hill should live surrounded by transmitters as I am. Radio 4 appears as 150mV of RF at the aerial and can be measured using a 20,000 $\Omega$  voltmeter in a tuned circuit. The poor transistor, be it unipolar or bipolar, just cannot stand the strain unlike the good old valve, which can be abused to a great degree before it gives up the ghost. Transistors and I.C.s die at the first spike or serious overload, but they still have their uses.

My final comments are that the valve will not be replaced for many years, and to this day there are jobs where the valve remains supreme master. Two examples are front-end stages of communications receivers and high power transmitters. Ask the Radio Amateurs! G. Sims (Derbyshire).

#### **Powerful argument**

With reference to Mr. Nuttall's comments in the November issue on your August Editorial, he complains of ignorance on your part, but what hypocrisy!

I refer Mr. Nuttall in the following to "Hi-Fi in the Home", by John Crabbe. (Never heard of it, Mr. Nuttall?)

The amount of power required by an amplifier in excess of 30W RMS has got very little to do with headroom. It is a question of relating power to loudspeaker efficiency, equated to the volume of the listening environment. Up to 30W power should give plenty of headroom. If it doesn't, then turning down the volume on any amp will assist, and if the result is then lack of loudness and realism in the music, this is a sign that the amp-to speakerto room conversion efficiency, is unbalanced. High power amps have uses, such as filling halls and auditoriums, but the myth that they are necessary to smooth reproduction all the time is a half-baked truth. Certainly they can help, but only when considering numerous other factors, such as the power supply, pre-amp overload, and speakers.

For the sound powers needed in all but the most exceptional domestic surroundings, the electrical power required from the amp depends mainly on speaker efficiency:- "The quantities of acoustic energy transmitted into the air in a normal sized room when reproducing orchestral music at a realistic loudness, are absolutely minute; a power level of about half of one acoustic watt representing the very maximum needed during peaks of the most enormous climaxes. But most loudspeakers are so inefficient that production of this 12W of sound output requires about 10W of electrical input. An electroacoustic efficiency of 5 per cent".

I can assure Mr. Nuttall that low power amps are quite capable of Hi-Fi. If they are not, then it is not the power output availability which is failing, but rather pre-amp overload, transistor junction storage capability, rise-time and slow rate related to negative feed-back. A 20W RMS amp with the correct degree of engineering will probably knock up 80W music power, which covers transient peaks quite effectively, other factors allowing. How the loudspeaker interprets this is another story again, depending on the thermal energy and general elasticity of the system amongst other factors.

The newer loudspeaker enclosures on the market are rarely above 2-3 per cent efficient throughout the whole audio bandwidth, so higher power amps are required. But not because they give more headroom, but because the speakers need on average a slightly higher drive level on

. music content, all the time. Check the difference between old and new speakers.

So, Mr. Nuttall, although at certain peaks, muddiness does set in on Hi-Fi systems (in this you are right). You are not strictly correct in interpreting its cause. Lets face it Mr. Nuttall, any amp (even your Naim 160) of any power description, will cause its output transistor to bang its head against the power rail and say, "that's enough, no more" if driven hard enough. It's only good preamp engineering, transistor stage design and your hand on the volume pot which prevents thisand here we reach the crux, which was partly the point made in the August Editorial. Power transistors cost many pence, signal transistors (even low-noise ones) are comparatively cheap. This is why it is possible for a well designed amp of £30 to be well up in Hi-Fi quality.

My point is made Mr. Nuttall. Your view on power reminds me of the gentleman in the sports car who once said to me, he believed he went a lot, lot faster with the hood down!

Wake-up Mr. Nuttall, and when you do, try appraising what you read or hear a little, before criticising.—E. Hoskins, (Technical Director, CEH Audio-Visual Enterprises, Tonbridge).

#### Any ideas?

Being an "old timer" I have been a reader of Practical Wireless for many years and am particularly interested in the "Going Back" series. A hobby (or rather a hobby within a hobby) of mine is the rebuilding of pre-war and pre-1930 radio designs and I have accumulated a large amount of gear of this date. One snag is that loudspeaker and headphone magnets have nearly always seriously deteriorated in strength and despite the most thorough reading of technical books available, no worth while information is available on the remagnetisation of these components (I think mostly of balanced armature speakers). Do any PW readers know of literature that deals with this, or can supply any info. on this subject.—T. Howells, (Radio Officer, m/v Geestcrest, c/o Geest Industries, Barry Dock, Glamorgan.

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160V	15p	LM556				22/6V		4700/40		100 for a	£26	250K Lin	30p
		LM710		DIL	50p		5p	4700/63		555 TII		500K Lin	30p
BZY88		LM273 LM723		DII		22/16 22/25	5p	POLY'	<u> </u>			1 Meg Lin	30p
400M.W	.	LM741			75p	22/25	5p 5p			100 for :	£3/	2 Meg Lin	30p
Zeners		LM741				22/50	6p	Mullard		RELAY	76	5K Log	30p
OV7	10p	LM747	14P	DIL		33/6V3	5p	Erie	3 01		C.O. H.D.	10K Log 25K Log	30 p
2V4	10p	LM748	8P I	DIL		33/16	6p	μ/F			Plug 1N	50K Log	
2V7	10p	NE555			45p	33/25	6p	-001	5p		£1 00 Each	100K Log	30p
3V0	10p	ME556		DIL		33/40	6p	.0022	5p			250K Log	30p
3V3		TAA30	ж		150p	33/50	7p	0033	5p	ASSO	RTED	500K Log	30p
3V6	100	TA A5				47/10 47/16	6p	·0047 ·0068	5p	POLYE		1 Meg Log	30p
3V9	100	TAAS	70 70			47/25	6p	-0068	5p 5p		CITORS	2 Meg Log	30 p
4V7	10p	TAA70	00			47/35	6 p	-022	Sp Sp	100 for a	E.S · 00		
5V1	180	TAD10	00			47/50	7p	∙033	5p	BVVA	DIODEC	ROTARY	
5V6	44.	T A D11	0		290p	100/10	6p	-047	5p		DIODES	SWITCHES	
6V2	10p	IIBA12	:0		125p	100/16	6p	11.	6p	1250 PI	V1 Amp.	BY	
6V8	10p	TDAM	, OC			100/25		-22	7p	100 10F	E-0 - 00	LORLIN	
7V5	10p	TBA56	NC NC			100/50 100/63		·33	9p	MULLA	PD		
8V2	10p	TBA72	0			125/15			20 p	4500/64	/ C 420	1P 12W	40p
9V1	10p	TBA99	000			220/16			25p	Screy	r C.432 Ferminal	2P 6W	40p
10V	10n	TCA 2	700			220/25	12p	4.7	35p	Joc. ew	75p	3P 4W	40p
11 V	40-	TCA27	ne '			220/50		6-8	40 p			4P 3W	40 p

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LA1466
LA1467 LA2330 LA2336 LA2400 LA2401 LA2402 LA2406 LA2501 LA2501 LA2502 LA2506 LA2512 VA1099 VA1106 VA1107 VA1110 VA1111 VA2600 VA3210 VA3410 VA3420 VA3430 VA3430 LA2517 LA2519 LA2532 LA2550 LA2554 LA2760 LA2701 LA2701 LA2705 LA2750 LA2755 V A 3708 VA8604 DIL SOCKETS 8 Pin 13p LA2754 LA2760 LA2761 LA2802 LA2852 LA2954 LA3105 LA3105 LA3127 LA3128 LA3145 LA3244 LA3305 16 Pin 15p LEDS RED -125 15p RED ·2 **DL747** 180p REED INSERTS FX1888 FX1889 FX2027 FX2142 FX3547 Lenth 1" 10 for 50p 100 for 350p 1000 for £25 FX2144 FX2211 LA3406 LA3441 FX2256 FX2270 FX2277 FX2318 FX2323 LA3445 I A3547 LA3504 LA3517

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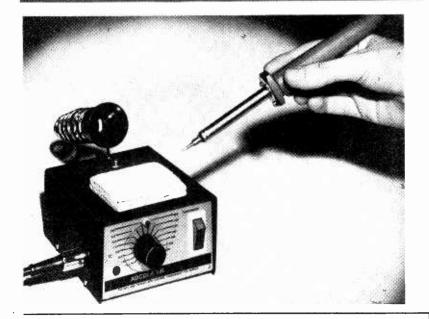
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### PRODUCTION LINES bill tull



#### Spot-on iron

I suppose there must be literally hundreds of soldering irons of one kind or another available in the UK at this very moment. Of course the cheap iron that is thrust upon an unsuspecting constructor, accompanied with the words 'psst-wanna buy a cheap iron guv' is little better than useless. A good maxim when looking for an iron is the one that says 'you pays yer money and takes yer choice', and although the latest iron from Adcola is a little on the expensive side, it could be argued that is is simply one of the best irons available to the general public.

Incorporating a thermocouple at the rear of the soldering bit, which then feeds an Op. amp and other circuitry, the temperature can be kept within  $\pm 2\%$  of the indicated dial temperature which is 120°C to 420°C. Heating time from a cool 24°C up to a blistering 420°C is claimed to be less than 2 minutes, while at any temperature the handle stays cool and comfortable for the user.

Operating from a 24V torodial transformer, this 50W, Unit 101 soldering station carries a price tag of £36.75.

Further information from Adcola Products Ltd., Adcola House, Gauden Road, London, SW4.

#### Space age tape

'Enough to go to the Moon and back 24 times' is the claim put forward by AMSA Magnetics, who further claim to be the largest manufacturer of cassette tape in the world. Realising the enormous potential of the UK market, they have launched a new range of cassettes which are said to be of a very high quality and feature a specially developed jamproof mechanism-Paraflo guides.

The range of cassettes available are Standard-a quality low noise tape with a permanent head cleaner leader, Extra-available in C66 or C99 lengths and featuring the new Paraflo guides and head cleaner Leader. Super-Ferric Oxide tape, and XHEthe top of the range tape containing Magnalinc<sup>TM</sup> Ferric Oxide which is claimed to give a superior S/N ratio and good frequency response. The XHE cassette also features instantstart recording and a jamfree mechanism. Prices range from 89p for a standard C-60 to £2.65 for a

AMSA Magnetics, 121 Loverock Road, Reading, Berks.

C-120 XHE.

#### Keep-it-kleen

Record cleaners tend to follow much the same lines these days, with the designs based loosely on the pick-up arm. In an endeavour to get away from this standard design, Bib have introduced a new parallel tracking record cleaner, called the Groov-Kleen 101.

As the name of this new cleaner

implies, a lightweight fitted with a velvet pad and cleaning brush, automatically tracks parallel across the record, silently, at the same speed as the pick-up arm.

The Groov-Kleen is supplied with



cleaning brush, a spare tracking brush plus velvet pad, together with spare self-adhesive base pads.

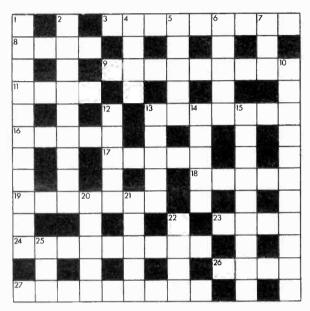
Price of this new Bib accessory is £3.97 and can be obtained from all good audio shops.

Bib HiFi Accessories, PO Box 78, Hemel Hempstead, Herts. Tel: 0442



Practical Wireless, March 1977

# DWTECHNICROSS No. 18



#### **ACROSS**

- Only one part of the set (9)
- A circuit-breaker misses a turn on it? (4)
- Type of receiver for dance or brass enthusiasts? 9 (9)
- 11 Answer distortion pique out east? (4)
- Repro blemish that needs solving inside? (7) 13
- Choose to add Ron to the energetic unit? (5)
- Turn in on cell starter for lightweight set? (5) 17
- 18 Does without a diode use to say goodbye? (5)
- 19 Lacking outside interference in pastoral areas (7)
- 23 Neat oscillator with a heat generator? (4)
- Microphone with an illegal bar? (9)
- 26 Only a saint understands this circuit! (4)
- It gives you a measure of power (9)

#### **DOWN**

- How often are they very high? (11)
- Harshness in the reception area? (9)
- The load we are carrying? (4)
- Oriental in part assembly for Dad? (5)
- Not a shilling Scottish audio for an Indian? (5)
- Two-way system for a dedicated lady? (3)
- 10 Equipment said to make you oldfashioned anon? (11)
- 12 Tax for a broken coral circuit? (5)
- Fear copper casing of lan's rectifier? (5)
- A lot of water from one A.C. vibrator? (5) 14
- 15 Live modulator at hand endlessly in a big vessel
- 20 Swapped tales of miniaturization? (5)
- Mrs Forsyte in the wire network? (5)
- Coast-to-coast hook-up excludes it? (4)
- He's no ham! (3)

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Do you wonder what nanofarad or nF means? It's all explained on the PW Info-Card for Capacitors and Resistors.



# PRACTICAL WIRELESS



If Part 1 of this project on page 940 has excited your interest then make sure that you order the April issue which will contain details of the remaining PCB's plus assembly and alignment information.

### TUG-O-WAR

It's usually a game reserved solely for those muscular, hairy he-men, but now you ordinary mortals have a chance to take part in PW's latest electronic game—Tug 'O' War.

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## SMOKE SENSOR ALARM

This project incorporates a sensor unit sensitive to many different gases and to smoke. A sensitivity control sets the level at which the audio alarm is triggered. Can be run from a 12 volt battery or the mains.

ON SALE IN MARCH

HILST touch selection of preset stations is not uncommon on modern TV sets, such a method of tuning has yet to make an impact in the FM tuner market. The tuner to be described is therefore unusual in that it possesses just this feature. Any one of eight pre-tuned stations may be selected simply by touching the appropriate "button" on the front panel. The tuner also has stereo defeat, mute and automatic frequency control functions and the required combination of these is set up for each station by a pre-arrangement of diodes inserted in two dual-in-line sockets. Thus when a station is



selected, the desired associated set of functions is also switched in.

The tuner itself makes use of fairly orthodox circuitry, using for the most part readily available integrated circuits and a Mullard tunerhead. Light emitting diodes are used to give indications of the station selected, tuning accuracy and signal strength, the presence of a stereo transmission and whether the AFC circuitry is operative or not.

Fig. 1 shows the power supply, which uses two 723 IC regulators. The main power supply rail should be set to about 12V. The tuner will work with anything between about 9 and 14V, but for a variety of reasons, the figure quoted is the most suitable. When a strong signal is being received, and consequently several LEDs are on, current consumption rises to about 200mA; under no-signal conditions it is only 70mA. The tuning voltage is obtained from a separate regulator, IC2. The touch selection ICs require a tuning supply at least 2V higher than the main voltage, and a value of about 17.5V is chosen. Both supplies include current limiting protection and the main power supply uses a power transistor, Tr1,

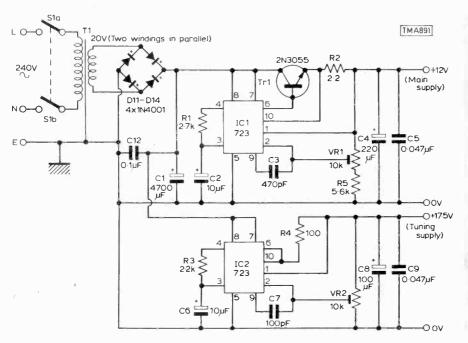


Fig. 1: The power supply section of the tuner provides two regulated voltages, one for the tuning circultry only and the other for the remainder of the tuner requirements.

Fig. 2: Circuit of the tuner proper comprising the tunerhead, IF amplifier IC3, demodulator IC4, stereo decoder IC5 and the audio output filter.



to dissipate the heat generated in dropping the voltage.

#### TUNER-IF-AFC

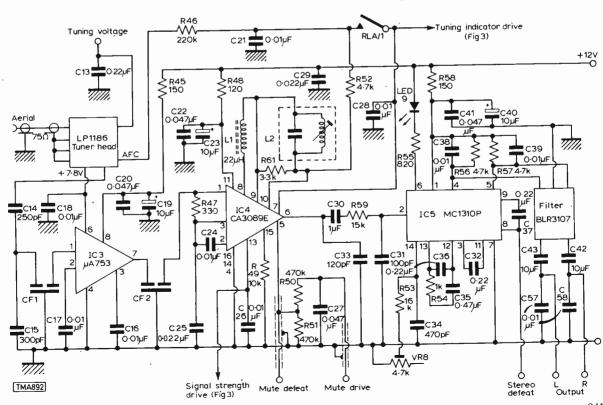
The circuitry for the radio part of the tuner is shown in Fig. 2. The tunerhead is a Mullard LP1186 module which is prealigned and provides a partially filtered IF output of 10.7MHz. This avoids one of the

most difficult problems of FM tuner construction, namely that of designing a good, trouble-free RF amplifier and frequency changer. IC3 is an IF amplifier designed particularly for use with ceramic filters (CF1 and CF2) in the configuration shown, having suitable matching impedances of  $330\Omega$ . It has two outputs of differing levels, and here it is quite sufficient to use the lower one (pin 7). The device has a useful additional feature in that pin 6 provides a stabilised DC voltage of 7.8V suitable for driving the tunerhead.

In the prototype CFSA10 7 ceramic filters were used; if however it is intended to use the tuner for long range reception where it may be rather trying to separate stations less than about 300kHz apart it may be worthwhile paying a little more for Vernitron FM4 filters. These have a better out-of-band performance and will make the receiver less susceptible to heterodyning of adjacent station signals (which have leaked through the system) with any stereo pilot tones present, producing annoy-

ing chirping in the output.

The well-known CA3089E (IC4) is employed to demodulate the filtered IF signal and a single-tuned detector coil (L2) simplifies alignment of this stage. The IC provides an AFC voltage on pin 7. When the AFC is switched on this voltage is fed via reed relay contact RLA1 to the local oscillator in the tunerhead. When off, the oscillator is left tied only to the reference voltage on pin 10 via R52. The tuning indication voltage is also obtained from pin 7, and a signal strength indication from pin 13. The IC also includes a muting circuit which cuts off the audio output in the absence of a signal of sufficient strength. In this instance it is not possible to drive the mute logic (input on pin 5) from the internal



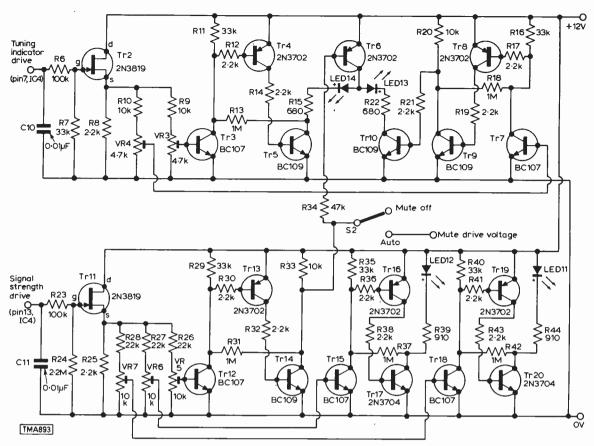


Fig. 3: Tuning indicator circuit including the LED tuning indicators is at the top with the signal strength and associated LEDs at the bottom.

mute drive voltage (pin 12) because the stability of this voltage is upset by the use of a high gain IF stage preceding IC4. Pin 5 is therefore driven by an externally derived voltage.

#### STEREO DECODER

The stereo output from IC4 (in its multiplexed form) is fed to the stereo decoder MC1310P (IC5) via C33, C30, R59 and C31 which produce a high frequency roll-off to help prevent leakage of adjacent station signals to the decoder (as mentioned before). This filter is not chosen for a -3dB point at 55kHz (the ideal situation) as this tends to diminish the stereo separation. More complicated circuitry would be needed to prevent the leakage and still maintain the full 53kHz bandwidth occupied by the two channels.

The decoder is a phase-locked loop system and the only adjustment required is alteration of the internal oscillator by VR8 until the device captures and demultiplexes the incoming signal. The decoding action can be defeated by earthing pin 8, and the means of doing this will be shown later. The output is fed to a DIN socket via the filter module BLR3107. This unit severely attenuates the 19kHz and 38kHz pilot and multiplex frequencies in the output. If this were not done the frequencies might cause disastrous whistling and howling when recording from the tuner by mixing with the tape head bias frequency. The presence of a stereo signal is indicated by the illumination of the beacon, LED9. Capacitors C57 and C58 are inserted to balance to audio output. The filter

should be terminated with  $4.7k\Omega$  resistors but a better tonal quality was obtained with the arrangement shown. The reader may wish to alter this to suit himself or his amplifier!

#### TUNING INDICATORS

A pair of LEDs is used to indicate the tuning accuracy. When properly adjusted both lights will come on, and as the receiver is de-tuned one way or the other, will be extinguished depending upon the direction of de-tuning. Similarly the signal strength is indicated by a pair of LEDs. Under good signal conditions both will be on, but only one or neither for weaker signals. With zero or extremely weak signal reception the tuning LEDs will also be extinguished. Most LED tuning indicators of the type described use a long-tailed pair, but the switching of the lamps by this circuitry is not particularly sharp.

The author preferred indicators with a sharp on/ off action, and to this end a voltage level detector circuit was designed with low hysteresis but a sharp transition. The basic unit is used five times in the indicator circuitry, Fig. 3. Each is fed by one of two FET buffers via a preset potentiometer (VR3-VR7) which determines the switching level of the circuit. Circuit action is fairly straightforward, the  $1M\Omega$  resistors introducing a small amount of positive feedback to sharpen the switching action.

The top half of the circuit shows the tuning indicator drive. It will be seen that the second level detector contains an extra transistor, Tr10. This

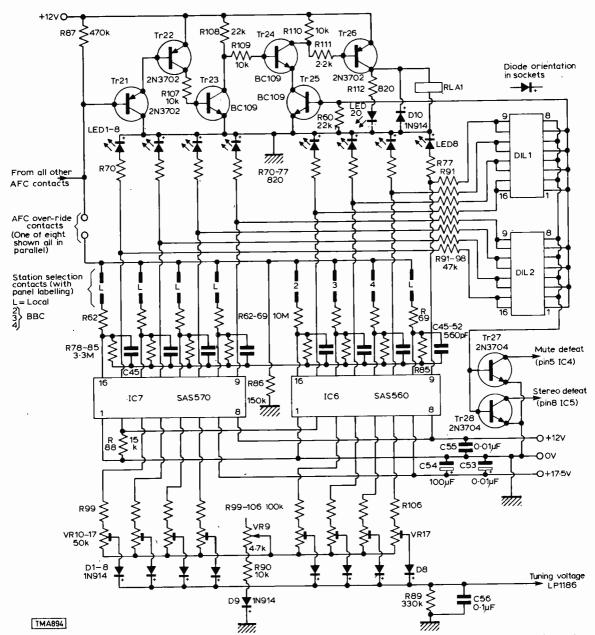
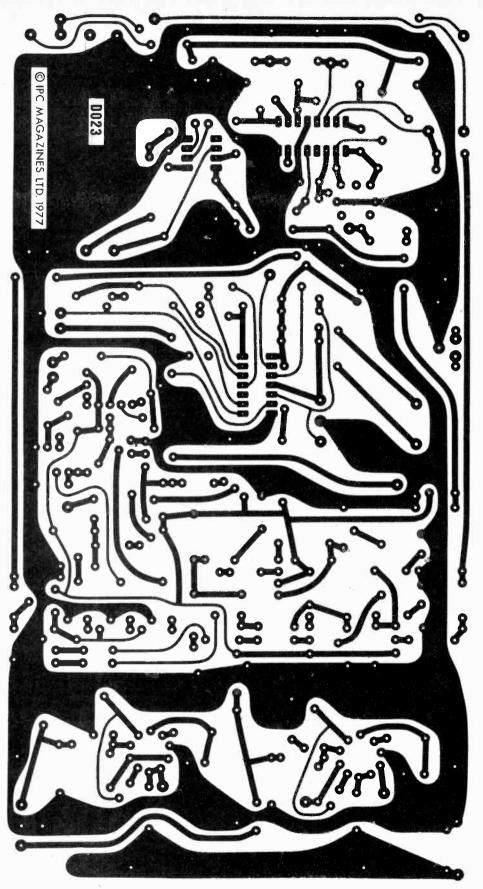


Fig. 4: All the circuits of Figs. 1, 2 and 3 are mounted on the main PCB. The eight touch buttons and the touch selection circuit shown above are mounted on two smaller boards mounted behind the front panel.

inverts the action of that circuit so that the system as a whole acts as a window voltage detector, providing indications as to whether the tuning indicator drive voltage from pin 7 on IC4 is above, below, or in the window. Only in the latter state is the tuner correctly tuned to a station. Tr6 is driven by a signal strength level detector from the bottom half of the circuit so that when that detector is switched off, Tr6 is off and both tuning LEDs are extinguished. This circuit also provides the mute drive voltage mentioned earlier. When Tr14 is on the mute drive falls to about 1V and the audio is not attenuated. S2 is fitted to the rear of the tuner chassis and may be switched either to auto (when the mute action is as described) or to "mute off" when the audio output is never attenuated, regardless of the signal conditions. The other two level detectors in the bottom half of Fig. 3 provide the signal strength drive to LEDs 11 and 12.

#### TOUCH SELECTION

All the circuitry so far described is mounted on PCB1 (Figs. 5a and 5b) which lies flat in the bottom of the chassis. Two more PCBs house the touch selection circuitry and they are mounted vertically, just behind the front panel and about <sup>1</sup>4in. apart. PCB3 is nearer the panel and on this are etched the eight touch "buttons", PCB2, behind PCB3, houses all the circuitry shown in Fig. 4, namely the touch selection ICs, the tuning presets and the AFC override circuitry.















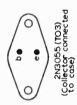


Fig. 5a: The foil layout for the main board PCB1 is shown here full size together with the leadout identification for the transistors, voltage regulators and LEDs.

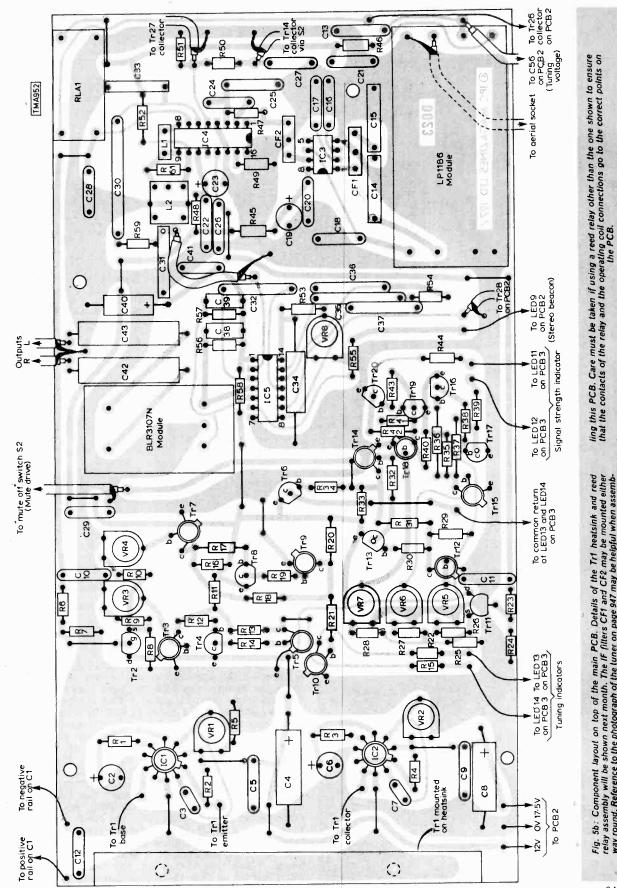


Fig. 5b.: Component layout on top of the main PCB. Details of the Tr1 heatsink and reed relay assembly will be shown next month. The IF filters CF1 and CF2 may be mounted either way round. Reference to the photograph of the tuner on page 947 may be helpful when assemb-

Practical Wireless, March 1977

Resistors			
R1 2·7kΩ	R27 22kΩ	R53	16kΩ
R2 2·2Ω(½W)	R28 22kΩ		tkΩ
R3 2·2kΩ	R29 33kΩ	R55	
R4 100Ω	R30 2·2kΩ	R56	
R5 5·6kΩ	R31 1MΩ	R57	
R6 100kΩ	R32 2·2kΩ	R58	
R7 33kΩ	R33 10kΩ	R59	
R8 2·2kΩ	R34 47kΩ	R60	
R9 10kΩ	R35 33kΩ		3·3kΩ
R10 10kΩ	R36 2·2kΩ		R69 10MΩ
R11 33kΩ	R37 1MΩ		R77 820Ω
R12 2·2kΩ	R38 2·2kΩ		R85 3·3MΩ
R13 1MΩ	R39 910Ω		150kΩ
R14 2·2kΩ	R40 33kΩ		470kΩ
R15 680Ω	R41 2·2kΩ	R88	15kΩ
R16 33kΩ	R42 1MΩ	R89 :	
R17 2·2kΩ	R43 2·2kΩ	R90	10kΩ
R18 1MΩ	R44 910Ω	R91_	R98 47kΩ
R19 2·2kΩ	R45 150Ω		R106 100kΩ
R20 10kΩ	R46 220kΩ		(2%)
R21 2·2kΩ	R47 330Ω	R107	
R22 680Ω	R48 120Ω	R108	
R23 100kΩ	R49 10kΩ	R109	
R24 2·2MΩ	R50 470kΩ	R110	
R25 2·2kΩ	R51 470kΩ	R111	
R26 22kΩ	R52 4·7kΩ	R112	
All resistors ¼V otherwise.	v 5% carbon	except wner	e indicated
		VD0 4 750	tto con
VR1, VR2 10kΩ		VR9 4·7kΩ	
VR3, VR4 4 · 7ks		bon p	
pre-s		VR10	
VR5VR7 10ks			t trimpots
pre-s		all 50k	
VR8 4.7ks		Presets are	
pre-s	et	type, horizor ing.	itai mount-
A		9.	
Capacitors			
C1 4700µF 40V	C	27 0·047μF*	
C2 10µF 25V		28 0·01µF*	
C3 470pF cera		29 0·022μF*	
plate		30 1μF*	
C4 220µF 25V		31 100pF mic	a
C5 0.047µF*		32 0·22μF*	
C6 10µF 25V		33 120pF mic	
C7 100pF cera		34 470pF poly	styrene
plate		35 0·47μF*	
C8 100µF 25V		36 0 22µF*	
C9 0.047µF*		37 0·22/4F*	
C10 0·01µF*		38 0·01μF*	
C11 0.01µF*	C	39 0·01μF*	
C12 0·1µF*	C	40 10µF 25V	(axial
C13 0 · 22 µF*		leads)	
C14 250pF mica	C	41 0·047μF*	
C15 300pF mica	C	42 10μF nor	-polarised
C16 0·01µF*		25V	3.50
C17 0·01µF*	C	43 10μF non-	polarised
C18 0·01µF*	12,77110	25 V	
C19 10µF 25V	C	45—C52 560p	F ceramic
C20 0·047μF*		plate	
C21 0·01 <sup>μ</sup> F*		53 0·01μF*	
C22 0.047µF*		54 100μF 25V	
C23 10µF 25V		55 0·01μF*	
C24 0·01µF*		56 0·1μF*	
C25 0·022µF*	C	57 0·01μF*	

A set of three PCB's can be obtained from the PW Readers PCB Service, see ad. on page 956. Semiconductors are available from Chromasonic Electronics, A Marshall Ltd., Watford Electronics,

\* C280 polyester. C44 is omitted

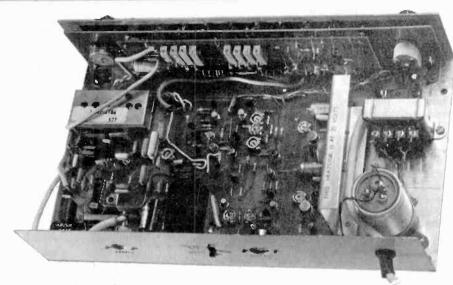
C58 0.01F\*

Semiconductors	
IC1 723 regulators	Tr15 BC107
IC2 (TO5 can)	Tr16 2N3702
IC3 µA753	Tr17 2N3704
IC4 CA3089E	Tr18 BC107
IC5 MC1310P	Tr19 2N3702
IC6 SASS60	Tr20 2N3704
IC6 SAS560 IC7 SAS570	Tr21 2N3702
ter encore	Tr22 2N3702
Tr1 2N3055	Tr23 BC109
Tr2 2N3819	Tr24 BC109
Tr3 BC107	Tr25 BC109
Tr4 2N3702	Tr26 2N3702
Tr5 BC109	Tr27 2N3704
Tr6 2N3702	Tr28 2N3704
Tr7 BC107	1120 2143704
Tr8 2N3702	D1 to D10 1N914
Tr9 BC109	D11 to D14 1N4001
Tr10 BC109	D11 to D14 1144001
Tr11 2N3819	LED1 to LED14 (GP,
Tr12 BC107	0.2" red, with clips)
Tr13 2N3702	Diodes for DIL
Tr14 BC109	Sockets:-1N914's
Miscellaneous	
Tunerhead, (Mullard	S2, Switch, SP on-off,
LP1186).	miniature slide.
L1, 22µH choke (Ambit)	Coaxial socket.
L2, 10.7MHz tuned cir-	DIN socket, 5-pin 180°.
cuit type KACSK-	Aluminium sheet,
586HM (Ambit)	18SWG
RLA Relay, reed type	300 x 180mm (11 <del>13</del> x
minlature, 12V 1kΩ,	7 <del>16</del> in.)
normally open.	300 x 90mm (11 13 x
Pilot filter, type BLR-	3½in.)
3107 (Ambit).	Mounting clip, vertical,
CF1, CF2, IF Filters	for C1.
CFSA10-7 (Ambit)	Mains cable and grom-
specify colour code	met.
as 'blank', or Verni-	Plywood and veneer
tron FM4 (Home	for sleeve.
Radio).	Rubber feet, 4 off.
T1, mains transformer	Panel knobs, 2 off.
240V/0-20V + 0-20V	Angle brackets for
150mA (6VA). (Dor-	sleeve, 6 off.
am 980-704).	DIL1, DIL2 16-pin IC
S1, Switch, DP mains	sockets.

Construction of the touch selection unit is simplified by the use of two ICs, SAS560 and SAS570, Fig. 4, which are purpose-built for this type of application. They have four high impedance inputs and provide the same number of tuning voltage and LED drive outputs. Any one station is selected by placing a finger on the appropriate pair of contacts, whereupon the tuning and LED voltages appear at the corresponding pins. By using a high tuning voltage (17.5V) and dropping resistors R99 to R106, a fine tune control VR9 on the front panel can be incorporated so that it gives approximately the same range of adjustment at the extremes of the FM broadcast band. Diodes D1 to D8 ensure that those tuning presets (VR10 to VR17) not being fed with the tuning supply voltage are isolated from the one which is. Each electronic "button" has three contacts; one is an earth link (via R86), one is connected to an input on IC6 or IC7 and one is connected to the base of Tr21. This transistor forms the input to a circuit which switches out the AFC (assuming it is on) when the contact is bridged to earth by a finger. When the AFC is on, the tuning indicator LEDs remain on over the whole range for which the

on-off, rotary.

C26 0.014F\*



General view Inside the finished Touch Tuner. The two PCBs behind the panel carry the touch buttons and selection circuits. The tunerhead is at top left with the audio output filter at bottom centre. At the right is the Tr1 heatsink, mains transformer and smoothing capacitor.

C	OLLATED	COMI	PONENT	SLIST	
Resistors	₹W 5% ca	rbon,	ex 2.2Ω		
2.20 1V		2.7kΩ	1	100kΩ 2	
100Ω	1	3.3kΩ	1	100kΩ	2
120Ω	1	4 · 7kΩ	3	150kΩ	1
150Ω	2	5.6kΩ	1	220kΩ	1
330Ω	1	10kΩ	9	330kΩ	1
680Ω	2	15kΩ	2	470kΩ	3
820Ω :	10	16kΩ	1	1ΜΩ	5
910Ω	2	22kΩ	5	2-2MΩ	1
1kΩ	1	33kΩ	6	3-3MΩ	8
2·2kΩ	15	47kΩ	9	10ΜΩ	8
Droset n	ots, minia	ture t	oriz.		
4 · 7kΩ		3	10kΩ		5
Pot. lines	ar carbon		4 · 7kΩ		1
	20 turn Cer		50kΩ		8
Timpot					100
Capacito	rs				
Mica	100pF	1	0.1	μF	2
	120pF	1	0.2	2µF	4
	250pF	1		7µF	1
	300pF	1	1 µ F		1
P'sty	470pF	1	Elec 10	F 25V	5
Cerami		1	10	F 25V	
	470pF	1		P	2
	560pF	8	100	μF 25V	2
C280	0.01µF	15		uF 25V	1
	0.022 uF	2		OuF 40V	1
	0.047µF	6			
Semicon	ductors				
IC's	723 (T05)	2	2N3	3819	2
10 3	µA753	1		107	5
	CA3089E	1	BC		2 5 7
	MC1310P				
	SAS560	1		0.2" Red	
	SAS570	1	V	Ith clip	14
Tr's	2N3055	1	Diodes		
11.5	2N3702	9		014	10+

station is pulled into tune. Thus to tune the station accurately the AFC must be temporarily defeated, and this is a convenient way of achieving it.

We now come to the two DIL sockets, Fig. 4. Conventional tuners have a row of switches on the front panel, controlling the action of such functions as the stereo defeat, the mute circuitry and the AFC. This tuner is exceptional in that the desired combination of these functions is preset for each station by the insertion of diodes in DIL1 and DIL2. Thus the desired arrangement is automatically switched in when the station is selected. So for Radio 2 (the station labelling is located between electrodes on Fig. 4) LED5 is illuminated; by connecting a diode across pins 1 and 16 on DIL 1 the AFC may be switched in for that station, providing base drive to Tr25 and to defeat the stereo and mute action a diode would be inserted across pins 2 and 15. Similarly diodes may be connected for other stations.

The mute and stereo defeat circuitry always act together since it is rare to require one without the other. Tr27 shorts pin 5 on IC4 to earth when turned on thus switching the mute circuit off, and likewise Tr28 shorts pin 8 on IC5 to earth which inhibits the decoding action of the stereo de-multiplexer IC.

It might be suggested that if tuning is all preset, there is no need for a mute circuit. However, when actually tuning, the interstation noise is very loud and best removed, in this case by the mute action. Secondly, when a station goes off the air that channel then is silent instead of producing 200mV of noise (unless of course the station is so weak that the mute had to be defeated to hear it).

IN PART 2, NEXT MONTH, WE DESCRIBE THE 'BUILD AND TEST AS-YOU-GO' PROCEDURE FOR CONSTRUCTING AND ALIGNING THE TOUCH TUNER. PLUS DETAILS OF THE BOARDS PCB2 AND PCB3 AND GENERAL ASSEMBLY INFORMATION

#### ON RECENT DEVELOPMENTS

#### The melting pot

Things are hotting up in the solar energy field. Latest method is the subject of patents and looks quite ingenious. With a solar cell, sunlight is converted direct into electrical energy. In this newer approach, the sunlight is converted into electricity by chemical means.

First step is to take a solar furnace (a suitable container with a dirty great lens focussing the sun's rays into the centre of it). Into this furnace is pumped some stuff called Sulphur Trioxide—chemically educated associates tell me that it's also known as SO<sub>4</sub>.

The Sulphur Trioxide gets hot and absorbs heat, and as it does so it separates by a process called dissociation. By pursuading the separated parts to pass through another chamber, they recombine to form Sulphur Trioxide again but in the process they give up the heat. The Sulphur Trioxide, now recombined and happy with life, is simply pumped back into the solar furnace for the next round of heat absorbtion and transfer.

Boom boom boom boom—SO<sub>3</sub>.

#### Ion bullets

Remember Jan's story about I2L, well I see that this new process involves a thing called ion implantation. To achieve this, the semiconductor substrate is prepared and set up inside a chamber. Then ions, tiny particles of the desired impurity, are 'fired' by an electronic machine gun at the substrate. If you were to look at the whole shooting match under a "times umpteen million" microscope, you'd see the lattice structure of the atoms in the substrate surface. And along, at great speed, would come these ions. Depending upon how you angle the substrate and thereby the lattice, will depend on wether the ions go straight into the lattice without striking any of the individual member atoms which make up this lattice, or wether the ions will collide with lattice atoms. If they do they will be slowed down and will not penetrate so far into the substrate. Thus

ion implantation is a very precise method of doping semiconductors and this process is used in the manufacture of this new l<sup>2</sup>L. Incidentally, the manufacturers are calling it "Substrate-fed logic" or Schottky l<sup>2</sup>L.

#### **Probing power**

New products are launched onto the market by their hundreds every month (some people argue that it's hundreds every week in electronics). Two which caught my eye last month might interest readers. One was a logic probe which allows one to prod round a logic circuit and find out what's happening. It's called the Catch-a-Pulse and will inform on six states; positive-going pulses, negative-going pulses, square waves, a logic one, logic nought, and no signal. Fountain-pen shaped it looked a useful device but most impressive was the price in the U.S.—just \$32.

The second device is a transistor: and what a transistor. It has the impressive name of STC9160 and the even more impressive power handling capability of 500W. This transistor is housed in a TO-3 package and is described by its manufacturer as a 50A device. It's quick, too. It will turn on in just 3.5 millionths of a second.

We've come a long way from those little "red-spot" transistors.

#### Small talk

If there ever was such a word as "Littlerer" it would surely apply to a new development at the research labs of a British manufacturer. Previously, I've mentioned a semiconductor technology called injection logic or I2L. What the British company has done is to adopt a significant variation of I2L with the result that the packing density-how many transistors can be got in a given size—has been increased some three times. At the latest head count they had managed to squeeze some 800 semiconductor gates into an area of 1mm x 1mm. Just get a ruler and look at one lone millimetre. Better still draw a 1mm square on a

piece of paper and then imagine 800 transistors inside that. See what I mean about "littlerer"?

#### Spot-on

Yet another interesting device mentioned at the Washington conference was a stable reference voltage source. This one is the brainchild of an American company and is still at the laboratory stage but it looks extremely promising. It employs a short channel MOSFET in what's called the punchthrough mode (don't even ask!) What is important is that when persuaded to supply a reference voltage at just over 3.8V, this device held the voltage stable to within +0.0005V. Even more important, it held the voltage to within that 0.0005V over the temperature range of 25°C to 50°C.

Perhaps the government could have a whip round and purchase one for the British economy?

#### **4-layer** switch

What about getting a Thermosenstor for Easter? I didn't believe it either but it cropped up at the International Electron Devices Meeting in Washington recently. It's basically a temperature-sensitive switching device and it's the handywork of a Japanèse company who gave a paper on it at the conference. The theory of it is quite horrendous (I reckon a degree in Venusian physics is required to understand it) but in simple terms it is a pnpn semiconductor, a four-layer device. If it is forward biased it can then switch on at a predetermined temperature. The precise turn-on temperature is decided by the value of a solitary resistor which is wired in parallel with the emitter of the Thermosenstor. The temperature switching range extends from 50°C up to 150°C.

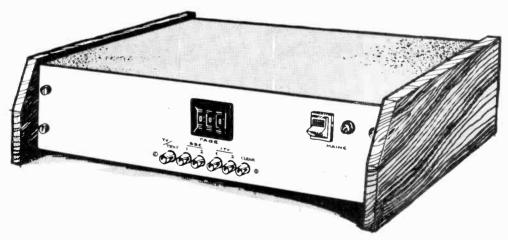
Ginsberg

# in next month'issue

# TELEUISION

MARCH issue on sale February 21st,

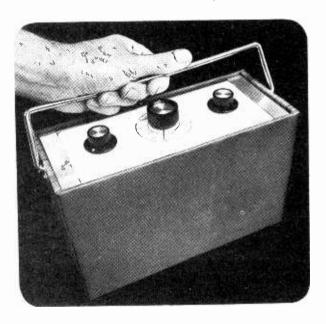
# The "TV" TELETEXT DECODER



Have you seen Ceefax/Oracle yet? By building the *Television* Teletext decoder you could soon be watching these services in your own home. It's designed to work with any 625-line receiver, either monochrome or colour, and has the advantage that you don't need to modify your domestic TV set in any way — just plug the decoder's output into the aerial socket. A Teletext decoder is complex, but to make assembly easy this design uses plug-in printed boards and a mother board, while the very latest receiver techniques are used in order to simplify alignment. The unit has its own mains power supply.

Also: SERVICING THE BEOVISION 2600 3000 3200 SERIES
FIELD TIMEBASE FAULTS ON THE RANK A823 CHASSIS
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# BASTYUU MARANER EGRAYER EGRAYER



THIS receiver uses three transistors and an audio IC in a circuit which is capable of very good results. With the usual form of receiver tuning, signals at the high frequency end of the medium wave band are crowded together and tuning is critical. As an aid to operation on these frequencies high-frequency bandspread is fitted, in addition to long wave and medium wave coverage. This provides three bands, approximately as follows:

Long Wave: 165 to 270kHz. Medium Wave: 550 to 1525kHz. Bandspread: 1475 to 1850kHz.

The receiver is not a miniaturised design which simplifies wiring and construction to some extent.

#### **Receiver Circuit**

The circuit is shown in Fig. 1 where L1 and L2 are the medium wave and long wave sections of the internal ferrite rod aerial. Base coupling to Tr1 is via L3 and a tapping on L2. Coil L4 is the oscillator coil, common to long and medium waves. For LW coverage L1 and L2 are in series, VC1 with integral trimmer TC1 being for aerial tuning. Switch S2 introduces C3 and TC4, loading the oscillator coil

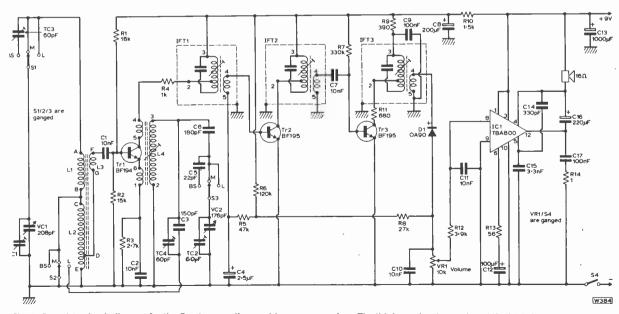


Fig. 1. Complete circuit diagram for the Easytune medium and long wave receiver. The third waveband spreads out the high frequency end of the medium wave band where congestion of stations is prevalent.

#### \* components list

4 WE 3 47250		* * * *
Resistors		
R1 18kΩ	R8	27kΩ
R2 15kΩ	R9	390Ω ·
	R10	
R4 1kΩ	R11	
R5 47kΩ	R12	
R6 120kΩ	R13	56Ω
R7 330kΩ	R14	1Ω
All 1 or 1W 5%		**
VR1/S4 10kΩ log pot w	vith swi	tch
		**************************************
Capacitors		
C1 10nF	C10	
C2 10nF	C11	10nF
C3 150pF	C12	100µF 6V
C4 2-5µF	C13	1000µF 10V
C5 22pF	C14	330pF
C6 180pF 5%	C15	3-3nF
C7 10nF	C18	
C8 200 uF 10V	C17	
C9 100nF	TC3/	
VC1/2 208/176pF, slow (Jackson 00)		
Semiconductors		
Tri BF194	D1	OA90
Tr2/3 BF195	IC1	TBA800
Inductances		
L1/2/3 MW/LW ferrite		
L4 Oscillator coll	Denco	TOC1)
IFT1/2 IF transformer	(Denco	IFT13) IFT3 (Denco
(FT14)		
Miscellaneous	,	
Speaker 160 3in. dia		
Bettery 9V and termina		
cabinet and covering, 10	$\times$ 4ln.	flanged plate (Home
		ber). Handle. Plastic
	ie made	a lizio a-bole a-way
feet. Two brackets, hom single wafer switch	ne made	a i jzjo o-pole o-way

for suitable coverage. This coil is tuned by VC2, TC2 being part of the ganged capacitor.

For MW reception S2 shorts out L2 and C3/TC4 are out of circuit. C6 is the oscillator coil padder. For bandspread reception, S1 places TC3 in series with VC1 and S3 places C5 in series with VC2. This results in a narrow frequency coverage for the full rotation of VC1/2. Switch S2 shorts out L2 as for ordinary MW tuning.

Silicon transistors are used for the self-oscillating mixer Tr1, and the two IF amplifiers Tr2 and Tr3. Diode D1 is the detector also providing an automatic gain control voltage for Tr2 through R8 and R5.

The required audio level is provided by the volume control VR1. The IC provides audio, driver and push-pull output stages, with diode compensation, and operates the loudspeaker directly. This audio IC has a low standing current and is economical while giving excellent volume from a 9V supply. S4 is the on/off switch incorporated in the audio gain control.

The three IF transformers are pre-tuned and there should be no particular difficulty in aligning the receiver even if no signal generator or other instrument is available.

#### **Circuit Board**

The board is approximately 127 x 70mm (5 x 2<sup>3</sup>4in), Fig. 2. The first step is to drill holes for TC4, L4 and the three IFTs. Position them so that the tags will come as shown in Fig. 3. A coloured spot marks pin 4 of L4. A <sup>3</sup>32in drill is suitable here, with a <sup>1</sup>16in drill for the can tags of L4 and the IFTs. No force should be used to fit these items and if necessary holes may be adjusted to give an easy fit by employing a small

It is as well to leave the transistors and IC1 off until later. Underneath wiring is shown in Fig. 3. Probably the simplest way is to begin by wiring the earth line to the can tags, using 24SWG tinned copper wire. Follow by placing resistors and capacitors as in Fig. 2, then turn the board over and shape and solder the wire ends, cutting off any excess leads. Sleeving is put on leads which cross or run near each other or near joints and tags. There is plenty of space for components but reasonable care is necessary to make a sound job here.

If necessary check resistor values as they are inserted. The electrolytic capacitors must be placed with the polarity shown. If each connection and component is marked in coloured pencil as completed it is unlikely that anything will be omitted.

#### **Semiconductors**

The pins of the transistors are placed as in Fig. 3 and leads are soldered to them. Avoid lengthy heating of these joints in particular. The iron should be at its full temperature and should be applied to the cored radio grade solder, which is held on the joint. The solder should flow in a second or so, and the iron is then removed at once.

Shape the ends of D1 to reach pin 4 and R8. Be sure this component is fitted with the polarity correct. Positive on the diode is indicated by a coloured band. If this diode is reversed the AGC action is upset.

The pins of IC1 are located as in Fig. 2 and spaced to fit the 0·15in matrix of holes. Bend the heat sink fins up slightly to clear the board and C14. Only pin 1 is marked on the actual IC. These pins come close together and they are small so a little care is needed when soldering them to connections, as in Fig. 3. Unused tags must not touch other connections or joints.

#### Wiring

Leads can be provided by fitting lengths of thin coloured flex or by inserting Veropins, which will take leads running to the appropriate points. A lead will run from C1 to F on L3, another from C3 and C6 to sections S2 and S3 of the bandswitch and a wire will run from 2 on the oscillator coil to the frame tag of VC1/2. In the audio section, D1 negative and an earthing lead from pin 5 of IFT2 run to VR1. Resistor R12 is soldered to the rotor tag of VR1 and is wired to C11 and pin 8 of IC1. Twin leads are provided from pins 3 and 4 of IC1 to go to the loudspeaker.

All these leads can be several inches long and arranged to come from the board as in Fig. 2 so that they can be readily identified later.

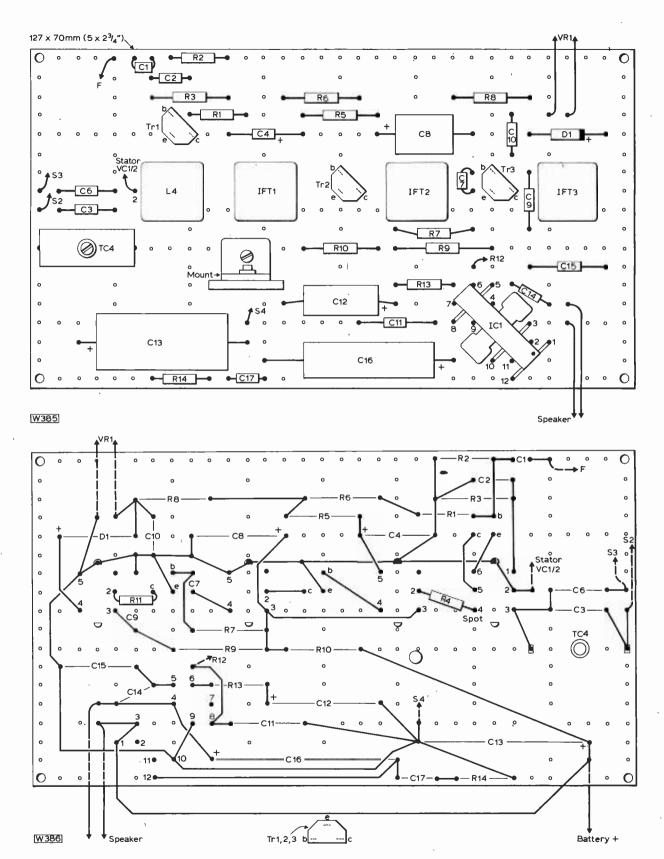
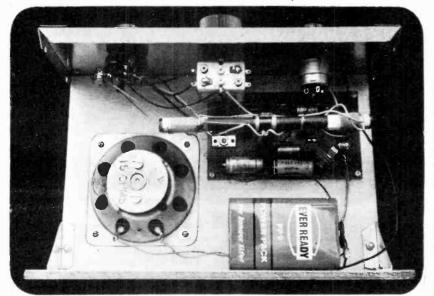


Fig. 2: at the top, shows the top of the veroboard and layout of components. The board size is 127 x 70mm (5 x 2½in.). Fig. 3: at the bottom, illustrates the wiring necessary under the board. In many cases the component lead is sufficiently long to make a connection.



General view of the finished receiver. The details of the bracket holding the ferrite rod aerial can be seen in Fig. 2.

#### Sub-assembly

This assembly allows the receiver to be made up as a complete working unit while all connections and the underside of the circuit board are easily reached. Cut a piece of 6mm plywood 250 x 160mm (9 $^{7}_{8}$  x 6 $^{1}_{4}$ in). At the right, 63mm (2 $^{1}_{2}$ in) from the right edge and 63mm (2 $^{1}_{2}$ in) from the bottom, mark a centre, and draw a 75mm (3in) diameter circle. Cut out this part to take the speaker. If a speaker of different size is used, cut this aperture to suit the dimensions of its cone.

Position the circuit board approximately 12mm ( $^12$ in) from the left edge and a similar distance from the top of the plywood. Mark round the circuit board with a pencil. This part is now cut out, but corners approximately 16mm ( $^58$  x  $^58$ in) and of tri-

angled brackets. This acts as a shelf for the battery and allows the receiver to stand upright when out of its case, or to rest on its back or either end. The top panel is a 254 x 100mm (10 x 4in) flanged "universal chassis member." Using a centre-line, punch a 16mm ( $^{5}_{8}$ in) hole in the middle for VC1/2. At 73mm ( $^{27}_{8}$ in) distance from this centre, punch holes for the bandswitch and VR1.

Place the top of the plywood inside the front flange and drill flange and wood for three countersunk 6BA bolts which will secure these parts together. VC1/2 is fitted by means of three countersunk 4BA bolts. This component is set back about 10mm ( $^{3}_{8}$ in) from the panel to bring the tuning knob lower (the spindle cannot be cut short due to the integral reduction drive). This requires spacers, or

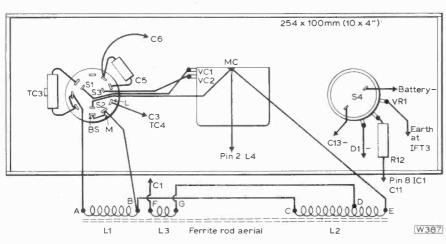


Fig. 4: Arrangement of the components on the front panel and wiring between the range switch and the ferrite rod aerial, and between the volume control and the board.

angular shape are left so that the circuit board can be secured by four woodscrews driven through the holes in Fig. 2, and into the wood. The circuit board is fitted with the IFTs towards the top and IFT3 at the edge. It is then possible to reach all connections, etc. under the board if this should prove to be necessary.

A piece of 6mm plywood 250 x 95mm ( $9^{7}_{8}$  x  $3^{3}_{4}$ in) is fitted to the bottom of the assembly by two right-

extra nuts, locked against the panel and capacitor. The ends of the bolts must on no account project beyond the thickness of the front capacitor plate.

Fig. 4 shows the panel and connections to the circuit board can now be fitted. The trimmers TC1 and TC2 are provided on the ganged capacitor, so are not shown separately. TC3 is soldered directly to the tags of S1 as shown. Capacitor C5 is mounted on S3.

#### **Ferrite Aerial**

The aerial can be mounted by cutting a strip of paxolin about 75mm (3in) long and securing it by a small bracket, Fig. 2. The rod is then bound to the top of the paxolin. An alternative method is to cut a piece of hardwood or insulating material about 63mm ( $^{2}$ <sub>2</sub>in) long from material about 12mm ( $^{1}$ <sub>2</sub>in) wide and 9mm ( $^{3}$ <sub>8</sub>in) thick. This can be tapped for a bolt to secure it to the board or drilled for a self-tapping screw. At the top, a 'V' is filed to take the rod which is held with adhesive and thread passed round the rod and through a hole in the mount.

Coils L1 and L3 are supplied wound on one former. Place this so that L3 is towards the middle of the rod, Fig. 4. Connect the beginning of L1, lead A, to S1 and wire F of L3 to C1. L2 is the largest winding, on a separate former. The tapping D is electrically near the end E. This can be seen by examination. Place L2 on the rod in such a way that the direction of rotation of the winding A to B is followed on by C to E. If this is not done LW alignment is impossible and L2 must be taken off the rod and placed the other way.

Point G of L3 is then wired to tapping D. Points B and C are connected together and to S2. Point E of L2 goes to the frame, MC, of VC1/2. These leads should be long enough to allow a little movement of the coils on the rod. Connections are best made with quite thin wire and sleeving is put on all the leads.

#### **IF Alignment**

The IFTs are aligned by the maker and it is better not to touch these until proper results are obtained. They may then be adjusted a little as necessary, using a proper trimming tool, such as that available from the maker. A wedge-shaped blade is likely to break the cores.

To adjust the IFTs tune in a weak signal, with VR1 turned up near maximum, and rotate each core for best results. Only a small movement should be necessary. If equipment is available a signal generator may be used, and output observed with an output meter. This allows more precise adjustment than does aligning the IFTs by ear. But despite this, perfectly satisfactory results should be obtainable without instruments. Once the IFTs are aligned, leave them alone, as they do not need touching when trimming the aerial and oscillator circuits.

#### RF Alignment

Trimmers TC1 and TC2 should be very slightly unscrewed. Switch to MW. Check that the medium wave coverage is from approximately 550kHz to 1525kHz. If there is much error at the LF end of the band rotate the core of L4 to correct this. Should there be considerable error at the high frequency end of the band, adjust TC2 as necessary. Subsequent adjustments depends on the positions of L4 core and trimmer TC2. This means that if any change is made to the settings of these, later adjustments will be upset. So the correct positions for L4 core and TC2 must be found first, as shown by correct MW coverage, with the bandswitch in its central position.

Tune in a weak transmission near the LF end of

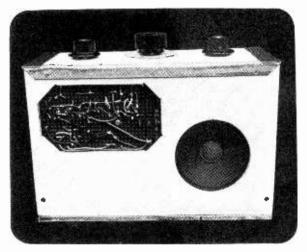
the MW band and move Ll on the rod for best volume. Then tune to a signal around 1500kHz and adjust TCl for best volume. As these adjustments influence each other, repeat them a few times.

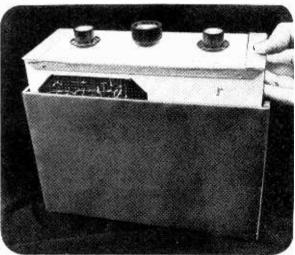
Switch to LW, set VC1/2 half closed and rotate TC4 until the 200kHz BBC Radio 2 is heard. Then tune to a signal near the LF end of this band, one or more foreign stations should be audible and move L2 on the rod for best results.

If these adjustments are difficult to achieve check wiring to L2 and its "sense" upon the rod, and also, if necessary, the setting of the core of L4.

#### **Bandspread Adjustment**

A signal should be tuned in around 1500kHz, VC1/2 nearly two-thirds closed, and TC3 should be rotated for best results. All the trimming adjustments can, of course, be made with the aid of a signal generator. When using actual transmissions for this purpose, avoid very strong signals. These will operate the ACG system and its action will mask adjustments so that a critical setting is not found easily. Distant signals will be subjected to





The top photograph shows the cutout in the base which enables the circuit to be worked on by merely pulling the set out of the cabinet. In the view at the bottom the set is lowered into its cabinet, when the speaker should align with the hole in the cabinet, shown in Fig. 5.

fading, especially on the higher frequencies, and this may make adjustment difficult. So choose a local transmission and carefully orientate the receiver for a low level of signal. Check again through each band to get best results.

#### Cabinet

Construction of the cabinet is shown in Fig. 5. The receiver panel or top is assumed to be 254 x 100mm (10 x 4in) so it is essential that the cabinet will clear this inside. Bearing this in mind the dimensions given are correct. The sides are of plywood while the front and back are hardboard. Glasspaper all meeting surfaces smooth and wipe away all dust. The sides are then secured to the bottom, using adhesive and a few panel pins. A 100mm (4in) diameter opening is cut in the front matching the

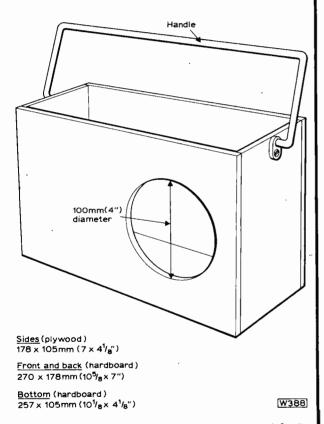


Fig. 5: Dimensions of the cabinet and material required for its construction.

position of the speaker. The sides and bottom edges and front are smeared with adhesive, placed together and fixed with pins. The case is then turned over and the back similarly fitted. When the adhesive is dry, the exterior can be sanded.

The back and front are covered with fabric secured with adhesive along the bottom of the case, at each end, and inside at the top, drawing the material taut. Any adhesive on the front of the case may penetrate the fabric and spoil the appearance.

Wood-grained self-adhesive material is then cut

to cover the sides and bottom. This is 100mm (4in) wide and overlaps into the top of the case. Four small feet are screwed on the bottom. The handle was pivoted in two small flat plates, screwed to the sides of the case. A flexible handle could be made from a length of material, folded and sewn, to match the case.

The receiver panel can be provided with a card scale, calibrated in frequencies with the aid of a signal generator, or with station names marked. A straight wire can be soldered to the spindle of VC1/2 or a dial can be cut from Perspex and fitted to the spindle.

The bare aluminium was covered with self-adhesive material and contrasting strips cemented on at each end. There is a great deal of latitude, here, for individual finishes to be used.

# KINDLY NOTE!

#### CHROMACHASE 4 December 76/January 77

Several readers have commented on the fact that the diode matrix in this project is unsatisfactory and many of them have suggested solutions. Although each solution appears to solve the problem each varies from the others.

However, common to all is the correction to D8. As it is used in the article it prevents any operation except Clock Reverse. The removal of this diode will enable the unit to function in all modes except Auto.

We are trying all the readers' recommendations as well as some of our own and will publish our findings as soon as they are clear.

#### OSCILLOSCOPE CALIBRATOR January 77

In the components list R12 should be  $100k\Omega$  and not  $100\Omega$ .

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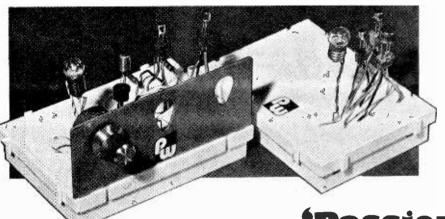
All boards, except SRBP, are glassfibre, drilled and roller-tinned

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		Polyphon, tune disc, blank, SRBI	P A008*	0 ⋅ 90 + 15
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# S-DeCnology



DAVID GIBSON



Passionometer<sup>9</sup>

HIS month's S-DeCnology circuit will be good news for those who have bought an S-DeC and have constructed any of the previous projects in this series. Good news because this project need not cost you a single penny more. All the components have been used in earlier projects so all you need to do is unplug them from your S-DeC and use them yet again.

With St. Valentines Day so close, it seemed a topical idea to build an emotion meter or "Passiono-

meter".

In practice, operation is very simple, with one person holding the end of a piece of wire, while a second person holds the end of another piece of wire. The couple can then hold hands, with their remaining free hands, and depending upon the "passion" involved, the bulb will light slowly from dim to full brilliance.

#### Super alpha configuration

Referring to the circuit in Fig. 1 the two BC109 transistors are wired in a super alpha configuration. This circuit arrangement gives a very high gain indeed. If one connected a resistor between the base and the positive rail of an ordinary single n-p-n transistor, its value could be quite high, the actual ohmic value depending upon the circuit con-

you will need . . . . .

Tr1 BC109
Tr2 BC109
L1 6V 100mA cycle bulb
B1 6V battery
One S-DeC.

figuration and other component values. However, because of the very high gain involved when using the two transistors as shown, the resistance between the base of Trl and the positive rail can be quite high and the transistor combination will still conduct sufficiently to pass enough current in Tr2 to light the bulb LP1. The numbered holes shown on the circuit diagram in Fig. 1 indicate the hole numbers on the S-DeC into which the particular lead is plugged. Almost any n-p-n transistors can be used for Trl and Tr2 but the BC109s were selected because they were already to hand from earlier projects in this series. The circuit works by using the body resistance of the two people involved. This body resistance has really been substituted for the base resistor referred to earlier. If you look at Fig. 1 you will see that holding the wires connected to

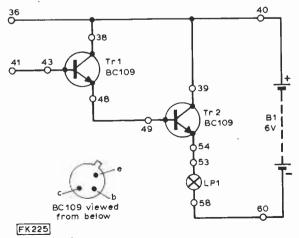
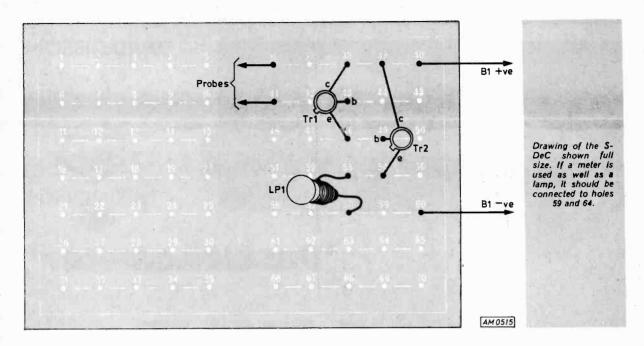


Fig. 1: Circuit diagram of this months project. All components can be taken from past S-DeC projects, which means no further expense.

S-DeC holes 36 and 41 will complete the base to positive rail circuit via the resistance of the bodies.

With fairly dry hands the bulb will probably not light at all. But if one waits for a few moments the fingers will perspire slightly and this "wetness" will

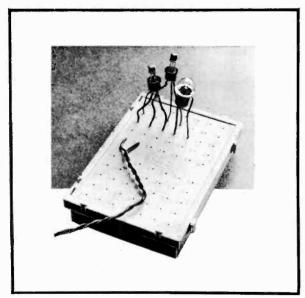
vincing version then you might consider purchasing a meter. It would be necessary to remove the dial and recalibrate it with segments labelled "Cold, Cool, Warm, Hot, Passionate and Danger". By simply connecting the meter in series with the bulb LP1



improve the conducting properties of the skin by lowering its resistance thus causing the bulb to start glowing.

#### Metered

As described, the "Passionometer" is very simple, cheap and will cause a lot of fun at any party. If you wish to make a more sophisticated and con-



Photograph showing the actual 'Passionometer'. As can be seen only three components are needed to complete this simple circuit.

it would read the current drawn by the bulb. Thus as the bulb started to glow brighter so the meter would read a higher "passion" score. A 6V 100mA bulb together with a 6V battery was found to be optimum, trading acceptable brilliance against reasonable circuit action.

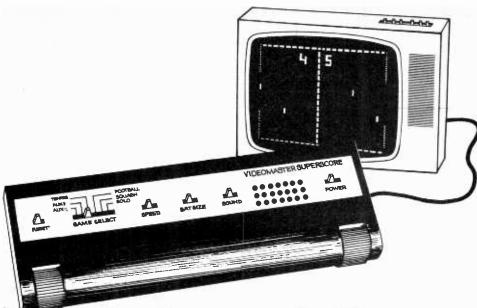
For those considering making the "meter version", the meter must be capable of reading 100mA FSD (full scale deflection). On test, simply plug the meter into S-DeC holes 59 (positive side of the meter) and 64, and transfer the battery negative lead from S-DeC hole 60 to hole 65. The probes were simply the free ends of two wires which were plugged into holes 41 and 36 on the S-DeC.

#### No guarantee

That's all there is to this months' circuit. Of course, I cannot guarantee that the device will indicate passion with any accuracy nor will it measure emotion—but you can have an awful lot of fun trying to prove me wrong!

Next months' S-DeC circuit again uses many of the components already employed in earlier projects. It can also be used at a party, or for more serious applications. Basically a miniature pocket stroboscope, it is simple and easy to build. It can be employed to give the stroboscopic motion-freezing effect commonly found at Disco's these days. It can also be used for more serious things such as splitting an apparently continuous stream of water from your tap into thousands of tiny crystal droplets.

The S-DeC project after that is even more fascinating . . . but I'll tell you about that in next months' Practical Wireless.



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#### by Eric Dowdeswell G4AR

LTHOUGH I haven't seen one yet I gather that we radio amateurs in the UK are to get a nice, new, shiny, and more expensive, licence from the Home Office as we renew our 'tickets'. Seems all the odd bits of paper separately authorising one to operate /T, /M and various odd modes not in the normal licence have all been abolished, and high time too. It is opportune to mention that the RSGB was fully consulted on the new licence at all stages and no doubt they have obtained every advantage for us that they could.

The three world regional organisations that make up the International Amateur Radio Union have reached agreement on the radio frequency allocations to be sought for the Amateur Service at the World Administrative Radio Conference to be held in 1979. Such early agreement augurs well for our cause but whether we amateurs get all that's being asked for will largely depend upon how well each national society presents its case to its own administration.

We should count ourselves as being very lucky to have the RSGB behind us considering the very high esteem in which it is held by our authorities but, regrettably, the opposite is the case in many other countries where commercial interests and politics decide such matters. Whatever happens at the WARC at least we shall know that the RSGB did its best on our behalf.

I wish it were possible for me to check back and see how long it takes someone writing in as a 'first timer' to go on and get his amateur radio licence. Jim Oram, just moved to a new QTH in Basingstoke, has been a SWL for about a year now and has joined the Basingstoke ARS. He hopes to take the RAE next May and while I am sure that we all wish him all the best I feel that he is being a trifle optimistic. However, have a go, Jim! The experience of taking the exam is worth its weight in gold, pass or fail. Jim has a Heathkit GR64 and expects to really get into the DX when he gets his new aerial up.

Alan Doherty BRS34968 is now in the peace and calm of Rotorua in New Zealand, at least until March next year when he is due to return to Northern Ireland. He has managed to listen around the bands

using two old valved radios, one as a BFO, so that he can copy SSB. Alan visits local ZL1QC who runs an FT401 and uses a Zepp aerial system, so he doesn't feel too home sick!

In Holyhead John Higginbotham now has a nice new FR100B which puts his old receiver to shame. He is also talking about an FL200 Tx and a linear, so obviously John is not too far off getting his RAE. Don't forget the Morse exam OM, you'll need it if you are going to use all that gear! The only QRM being suffered by John at the moment is that caused by exam worries! Newcomer Bob Harvey of Oxford is 15 and has got under way on the SW bands with a Codar Mini-Clipper but feels he could do with ambitious. Perhaps more Christmas came up trumps OM? Anyway, Bob added an audio amplifier and a large dial which he calibrated himself, and with a 32ft. aerial he started to hear things. Bob, too, has made contact with his local ARS and looks forward to starting an RAE course soon.

Anent the misfortunes of Robin Bayley of Allrighton whose local RAE course collapsed, he heard from G4FGR who put him on to another course so let's hope it all proves worthwhile for Robin. He now has an Eddystone EC10 which enables him to copy CW and SSB from 15 to 80m, using a long wire. Readers often refer to a 'long wire' aerial but I wonder if that is really what they mean. In this sense it should mean that the aerial is long in terms of the wavelength which would mean 275 odd feet to be a couple of wavelengths on the 40m band! I don't think that many readers can go to that sort of garden! Most aerials are much shorter which is why it is so essential to employ an aerial tuning unit to make up the length electrically.

In Harrogate Steve Cottis A8961 reports new countries with VP5, 5W1, VQ9 (Chagos Is) and FR7 all on 80m. He finds 20m closing down around dusk although the DX is still there after the Europeans have gone. Brian Harrison found something for the log on most bands down in Hastings, including rare KX6MU on 20m SSB. FO8EB is another that is quite rare and a worthwhile at any time on any band. Arnold Silverwood won't mind if I call him a 'first timer'. He lives in Nottingham and has built himself an HAC receiver and is starting to understand some of the amateur 'language'. While his set can be made to give quite good results I have pointed out to Arnold that there are more sophisticated sets around, and that he shouldn't write off our hobby too soon, if he feels he is not getting results!

A very interesting club magazine that comes my way each month is TARS TALK, the 'voice' of the Torbay ARS, edited by F. Bolton G3VTQ. Besides



all the usual club chatter there are a couple of small but interesting articles in the November issue on Roof Space Aerials and Receiver Drift. At present, meetings are held at Bath Lane, Torquay (rear of 94 Belgrave Road) but more suitable accommodation is being investigated. Anyone interested in joining the TARS is recommended to contact the Secretary, M. Yates G3UIQ at 23 Waverley Road, Newton Abbott. Devon.

In Hastings, Jesse Luxton G8GMI is not bothering too much about VHF but concentrating on getting his code up for the examination. Unfortunately Jesse had a stroke in 73 which he feels has not helped at all. He recommends PA0AA for slow morse as they run a course over the air and suggests getting their booklet from PO Box 2083, Eindhoven, Holland for 6 IRC's.

Andy Grendon of Dun Laoghaire, Co. Dublin has really got going since I wrote to him recently. He joined the IRTS and is now EI508 and attending Morse classes while a Heathkit SW717 helps him to cover the HF bands. He is in a built-up area and has little space for aerials. A 30ft end-fed suffices for the moment.

#### Log Extracts

R. Bayley:— 80m CW VS6DO ZL2BFU ZS2PR 40m CW CT2BP EA9FD K8CCV VE1AF 20m CR6LF FP8DH HK0COP TG9TI VP8AA 15m AP2AD DU1DBT VK3XI ZP5CD

A. Doherty:— In New Zealand 20m DL8NU EA3AEA EA6CK SV1BR

S. Cottis:— 80m A4XGB FP8DX (QSL K90TB) JA6YDG WA6EGL/VQ9 (Chagos Is) 20m CX7BV FP8ZZ FR7BI HK1CMX VP2GMB 5W1AB 15m A9XBD JY9CR VP5A YV2AMM 10m ZS1EZ

B. Harrison:— 80m ZC4FS HP1YV 7X2HM 20m DU2EL KX6MU 9D5A (Iran) 15m PY0ZAE (Trinidade Is)

All SSB except where indicated otherwise.



#### **MEDIUM WAVE DX**

by Charles Molloy

RANSATLANTIC DX dominates the scene this month, reception on this path being better than it has been for several years. Reader's logs show stations not normally heard while others, trying this path for the first time, are able to report success. Laurence Bennett writes from Bristol "at long last, mainly through help from your column, I have heard and verified my first transatlantic DX catches. Here is my complete North American log to date". Many thanks for the bouquet, Laurence. The

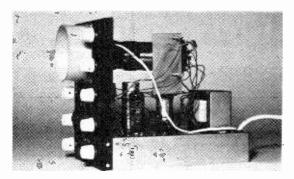
receiver is a Trio 9R59DS connected via an aerial tuning unit to a 68ft longwire aerial and the log consists of three stations from Newfoundland, CBN on 640 kHz, CKVO on 710, CJON on 930 plus four in New York City, WOR on 710, WCBS on 880, WINS on 1010 and WNEW on 1130. DXers who have still to hear North America should try after midnight for one of these stations, and if the path is open then at least one should be heard.

Laurence raises an interesting point when he says "I managed to hear both WOR and CKVO on 710 at the same time which resulted in a jumble which was difficult to resolve". DX signals on the MWs are usually subject to slow fading, the cycle often lasting a minute or two. This is typical of E layer reception which is quite different from the fast fading experienced on the short waves and, occasionally, at the high frequency end of the medium waves, where the F layer is the part of the ionosphere responsible for propagation. The technique on the medium waves is to stay tuned to a channel for several minutes when hopefully DX will appear out of the noise or, in the case of WOR and CKVO, either one will become dominant long enough for an identification to be heard.

Simon Knott who lives in Cambridge reports two instances of DX on a channel normally occupied by a North American "local". Using a Trio 9R59 and a 48in loop he heard CFRB in Toronto as well as WINS on 1010kHz and on 930kHz which is normally the domain of CJON he logged CFBC Saint John City in New Brunswick. CFBC sent a QSL after a month and Simon asks if this station has been logged by anyone else recently. CFBC is something of a rarity in the UK because of the directional aerial used at the transmitter. This produces a null towards the north east which protects CJON from interference within its service area but, unfortunately, DXers in Europe as well are prevented from receiving its signal. Simon refers to CBM Montreal on 940kHz which he hears occasionally. CBM has an omnidirectional aerial which presents a rather weak signal to any European DXer who is patient enough to sit on the channel when the path to North America is open.

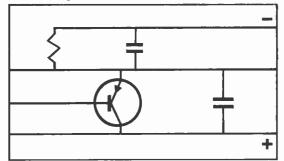
Derek Taylor (Preston) is still trying to identify a US station on 1390kHz. Although he taped the identification it sounds different every time he plays it! WCSC in Charleston, South Carolina, is a likely candidate here, the writer having heard it several times and the other possibility is WEGP Presque Isle in Maine. Derek who uses a Realistic DX160 and loop also reports hearing WJR Detroit on 760 at 0109, WHAS Louisville Kentucky on 840 together with All India Radio outlets in Bikaner on 1330 (at sign-on at 0100) and Gwalior on 1390 at 0125, Radio Antilles, Montserrat on 930 at midnight, Radio Victoria in Aruba NWI on 960 at 0040 and WMDD in Puerto Rico on 1480 at 0310. Puerto Rico, although in the Caribbean, is one of the states of the USA and consequently US callsigns are used. Surprisingly, there are 52 medium wave stations operating in Puerto Rico and since the language can be Spanish or English it is very confusing for the DXer. Other Puerto Ricans heard in the UK are WKAQ Radio on 550, WKVM in San Juan on 810, WHOA Radio Capital in Hato Rey on 870 and WUNO (Radio Uno) on 1320.

Steve Whitt (G8KDL) who DXes in London with a Chapman S6BS communications receiver, PW loop and differential amplifier (*Practical Wireless*, April 1973) has been experimenting. He replaced the



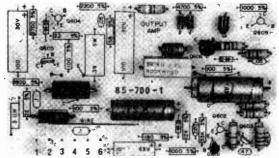
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2N3819 FETs in the differential amp with 2N4416s and the resistors with metal oxide types and obtained greater gain and less cross modulation. He says that conditions have been excellent and on some days (nights!) WINS on 1010 has been roaring in as early as 2230 causing interference to the Dutch station (Lopik) on 1007kHz. Steve finds North American stations excellent verifiers, he sends an International Reply Coupon with every report, and quotes from a letter from WBZ Boston (on 1030) which suggests that since they beam their signal on 270° he may have received them after their signal had circled the world! Unlikely, as propagation on the medium waves requires a path of darkness between transmitter and receiver.

More DX from Michael Illingworth of Christchurch who reports hearing KMOX in St Louis on 1120 with his Trio 9R59DS receiver. This is a really good catch which is only logged when reception is really good. Incidentally, US stations use one of two prefixes in their callsigns. Those beginning with the letter K are with one exception, KDKA, to the west of the Mississippi while those with the letter W are to the east. A callsign beginning with K means real DX from the UK.

An unusual request for identification comes from Nick Hall-Patch of Victoria, British Colombia, who, with a friend, heard an unidentified broadcast, in a transatlantic direction on 566kHz. It was at 0725 and consisted of a single short pip every 15 seconds. The station according to a law well known to medium wave DXers, faded out just before 0730 when of course Radio Eireann would have signed-on for the day. The writer had never noticed these pips before which only shows that one man's DX is another's QRM. (BBC Radio 4 also uses them—AED.)

Bob Bonsall of Buxton wants to buy a good radio that will pick up North American stations on the medium waves. Reports this month come from users of the Trio 9R59D/E, Realistic DX160 and Chapman S6BS. It would be a great help to readers starting the hobby if all logs were accompanied by full details of the receiver, aerials and any other gear in use. The time, in GMT, of any outstanding logging would also be invaluable.



### SHORT WAVE BROADCASTS by Derek Bell

THE Christmas season has slipped into history and one can collect together not only one's thoughts but the paperwork that floods in at this time! From Radio Havana there is the usual

collection of those awful calendars, the size of a matchbox, printed with the dates, unrelentingly, in Spanish! Radio Havana is now asking for letters to be sent to Apartado 7026, Havana, Cuba, and is sending out a schedule which advises that broadcasts for northern Europe will be on 17885 between 2010 and 2140, a very sparse service, with more time devoted to English language broadcasts to north, central and south America on 17750, 11865, 11760, 11725, 9685 and 9525. Nearly 16 hours, in fact!

Alongside this there is the usual crop of "Happy New Year" cards from other similar countries who seem to ignore the fact that the countries they send them to also celebrate Christmas!

Tony Cook, at present living in Gibraltar, logs a few of the signals that whizz around the ether over the Rock. These are as follows: Radio Togo on 5048 at 2100, Morocco on 4650 at 1930, Radio Prague (African Service) on 5760 and 5350 at 1900. Tony asks for information on the HAC Mk 2 set. I must confess I do not know a lot about this set but perhaps other HAC users can answer Tony on the problem of why the reception goes "up the spout" when he connects the bandspread! Finally, Tony would like to know of any books on the subject of DXing and here I can help. Radio Nederland has a very wide range of pamphlets on all aspects of the hobby and their address is PO Box 222, Hilversum, Holland. These publications are all free.

Lawrence Bennett, from Bristol, takes me to task for the lack of Latin American and African stations in the column. Regular readers will remember that a few months ago, with the aid of a Latin American reader resident in London, we looked at the LA scene. However I can only report on the logs that I receive and if no one logs LA then I can't print it! It must be said that it is a long time since anyone sent in such a comprehensive list as Lawrence has and here are a handful of the "DXotics" that his Trio 9R59DS has pulled in via a longwire:—

Radio Difusoras, Venezuela on 4890 at 2355
Radio Timbira, Brazil on 4975 at 0430
Radio Sutatenza on 5095 at 0045
Sierra Leone on 3316 at 2153
Radio Banqui on 5038 at 2200

Perhaps it's the thought of the very late nights that puts off the majority of DXers but it is a fact that Latin Americans are strictly for the birds, night owls that is.

M. G. Hayman has also a pet peeve and that is the fact that he waited for the arrival of his Codar Multiband 6 only to find that the firm had sent the order in two lots, and one parcel had got lost in the post! This must be a situation that drives the homebrew enthusiast right up the wall! Meanwhile he has to make do with his HAC Mk 2.

The final letter this month is from V. Marland who resides in Connahs Quay, Clwyd, who requests information on the way that a station report should be layed out. This is very simple, the station name first, then the frequency plus the time in GMT. It is not worth logging signal strength since this can vary enormously. Add some notes on the programme heard and any identification signal heard. Notes on interference to the station are sometimes helpful. If you enclose a couple of IRCs this will expedite a confirmation but don't be too impatient! For now I will wish you best 73s to you and yours.





#### by Ron Ham

THE sun has been very quiet during December; in fact the writer recorded only one burst of solar radio noise (midday December 8th) and that was the first since November 21st. Alan Taylor, G3DME, Crowborough, Sx, heard the "seashore sound," normally associated with solar noise, on the 10m band at 1747 on the 15th. Although the sun was set there could have been a large solar burst and Alan detected the radio signals as they came above the horizon. These reports are valuable and since Alan collects the International Beacon Project reports for the RSGB, any that you send to me will eventually end up with him.

At 0947 on Dec. 15th the writer heard the signal from the IBP Beacon in Cyprus (5B4CY 28·180MHz) and that was the first time for a month. An interesting observation came from Walter Butt GC2FZC, St Peter Port, Guernsey, who noticed that when he can hear DL0IGI and 5B4CY on 10m, conditions on the 2m band are above normal. What about it readers? The writer would like to hear a lot more about this.

Very soon the number of sunspots will increase as the new cycle gets underway, and Derek Poulter G3WHK, Morden, Surrey, a keen all-band DXer, wants to know how to observe these with a telescope. First of all Derek, NEVER look at the sun directly through a telescope or any other optical aid, it is dangerous to the eyes, in fact it is unwise to look at the sun at all. The technique is to point your telescope toward the sun and project its image on to a white card (shaded from the main sunlight) and pencil in the spots that you can see. Most basic astronomical books will show this.

Of course, you can always direct your 2m beam toward the setting or rising sun and listen around 143MHz (Quiet spot outside amateur band) for the "seashore sound" or "whooshing" which is the radio noise from a sunspot.

Although the atmospheric pressure was below  $30 \cdot 0''$  for most of December it did rise sharply from  $29 \cdot 5''$  at noon on the 8th to  $30 \cdot 0''$  by noon on the 9th and continued to rise slowly reaching  $30 \cdot 4''$  by noon on the 12th, and true to form, as this high pressure system moved southward, and the AP began to fall, there was a tropospheric opening.

During the late afternoon of the 14th, Roy Bannister, G8LXR, Lancing, Sussex, worked several French stations on 2m SSB and on the 15th Roy heard a PAO and an ON come up on two. From his QTH at Newhaven, Sussex, Alan Baker G8LGQ, worked into the Midlands during the evening of the 14th and also noticed patterns on his TV screen, another sure sign that an opening is in progress. From midday until late evening on the 15th the

writer was receiving a strong picture from Lichfield on Chan, 8 (189MHz) and around 2000 a 559 signal from the Sutton Coldfield beacon (GB3SC) on 70cm.

At 0930 on Dec. 5th, Joost Berden G3RND, Cowes, IOW, was taking his usual weather picture from the NOAA5 satellite as it passed over and was surprised to find that it's 137.50MHz signal was weak and fluttery for about five minutes, a long time in a pass. The satellite came in at 20°E for an overhead pass, but the signal did not become strong and steady until it was 65°N. At that time, Joost noted that conditions on 70cm were reasonable, he could hear GB3SC and during the previous evening he worked G3OSS in London for the first time, using SSB. There was definitely movement in the troposphere throughout both days, therefore it is possible that this region absorbed the satellite's VHF signal for a short time. The writer has known times when conditions were "up" on 70cm and flat on 2m.

Derek Poulter, G3WHK, has worked a lot of 2m DX by keeping an eye on the barometer and he makes a point of watching the late night weather map on BBC TV to see if there are any "highs" building up. Derek has moved QTH and just received planning permission for his new aerial array:—HQ1 mini-beam for 10/15/20, 4 element yagi for 70MHz, 2 x 6 element arrays for 2m, and something for 70cm, and all supported by a 40ft Versatower! Our readers will be looking for some good reports, Derek!

Another barometer watcher is that experienced VHF DXer, GC2FZC, who monitors the 2m beacons GB3CTC and GB3VHF daily and relates their signal strengths to the prevailing atmospheric pressure. He also has a regular Sunday morning sked with G3IEA and G8DOJ, both in Torquay, and they seldom miss out. In fact the sked with G3IEA has been going on for some 15 years!

Don't forget, if anything unusual occurs on the VHFs the writer will be pleased to have the details. After all, it is your observations that help to put a particular propagation event into its proper perspective.

Late Note: A slight change in the prevailing high pressure during the evening of 23rd December caused a predominantly N/S opening which upset UHF TV and provided some good DX on both 2m and 70cm. Between 1900 and 2200 the writer received 599 signals from beacons GB3SC (432·890MHz) and GB3EM (432·910MHz). At 2117 a 59 signal from G3DY in Northants, while on Ch. 8 (189MHz) a strong picture was received from the IBA transmitter at Lichfield. For most of the evening the 2m band was open from France through to northern-G.

#### **BROADCAST BANDS**

Short Wave Reports by the 15th of the month to Derek Bell, c/o Practical Wireless, Fleetway House, Farringdon Street, London, EC4A 4AD. Medium Wave Logs to Charles Molloy, 132 Segars Lane, Southport, PR8 3JG.

#### **AMATEUR BANDS**

Logs covering any amateur band/s in band/ alphabetical order by the 25th of the month to Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey, KT21 2TW.

#### VHF

Reports on VHF matters to Ron Ham, Faraday, Greyfriars, Storrington, Sussex RH20 4HE.

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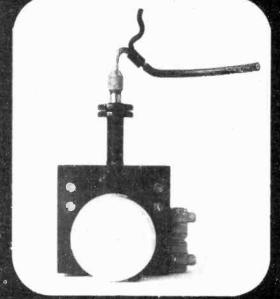
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SPECIFICATIONS
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OUTPUT POWER 60W RMS Into 8Ω LOAD IMPEDANCE 4-16Ω DISTORTION 0·04% at 60W at 1kHz SIGNAL/NOISE RATIO 90dB FREQUENCY RESPONSE 10Hz-45kHz — 3dB SUPPLY VOLTAGE ± 35 V SIZE 114 50 85mm

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APPLICATIONS: HI-FI-Disco-Monitor-Power slave-Industrial-Public Address

SPECIFICATIONS
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components. APPLICATIONS: Public address-Disco-Power slave-Industrial

SPECIFICATIONS
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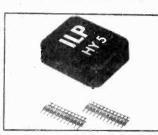
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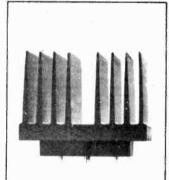
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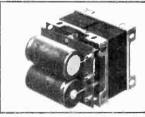
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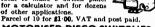
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ECF82 ECF86 ECH81	0 · 90 0 · 78 1 · 73	EY88 EZ80 EZ81 GY501	0 · 81 0 · 84 0 · 70 1 · 73	850 PD500 PFL200 PL36	2 · 62 1 · 23	UCL82 UCL83 UL84 UY85 6/30L2/ ECC804 6F23/EF81 30C1/PCF	1 02 1 02 1 23 0 73 1 20 2 1 16	PC801 30P19/ PC802	1:1:1:1:1:1:1:1:1:1:1:1:1:1:1:1:1:1:1:	06 35 47

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	0.47	EL95	0.60	PCL83	0.70	5Y3GT	0.65	12AX7 12BA6	0 · 38
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EBC41	0.75	EM85	1 - 00	PCL805/8		6AM5	2 00	30C15	1:00
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)(	ν.		4A		0.85	63
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î	T	EST	MET	<b>ERS</b>		
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ı	A	VO.	72			5 86
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50	0.20*	0.25	0.35	0.32	0-41	0.42	0:47	0.96
100	0.25*	0.25	0.40	0.37	0.47	0.48	0.54	1:02
200	0.27*	0.35	0.45	0.40	0.58	0.60	0.68	1:14
400	0.30*	0.40	0.50	0.45	0.87	0.88	0.98	1 · 40
600	_	0.65	0.70	_	1-09	1 - 19	1 26	1-80

#### TRIACS (PLASTIC TO-220 PKGE ISOLATED TAB)

										-
		4A	6	·5A	8	·5A	10	A	15	iA
	(a)	(b)	(a)	(b)	(a)	(b)	(a)	(b)	(a)	(b)
100V	0.60	0.60	0.70	0.70	0.78	0.78	0.83	0.83	1-01	1-01
I 200∨	0.64	0.64	0.75	0.75	0.87	0.87	0.87	0.87	1 - 17	1.17
400V	0.77	0.78	0.80	0.83	0.97	1 - 01	1 · 13	1:19	1 - 70	1 - 74
600V	0.96	0.88	0.87	1 · 01	1 · 21	1 .26	1 · 42	1.50	2.11	2.17

N.B. Triacs without Internal trigger diac are priced under column (a). Triacs with internal trigger diac are priced under column (b). When ordering please indicate clearly the type required,

#### **OPTOELECTRONICS**

Disp	lays	Discretes	
704	0.95	0·2 Red	0.13
707	0.95	0·2 Green	0.20
727	1 95	0.2 Clear	0.14
728	1 - 95	OCP71	0.90
747	1 · 85		
700	4 05		-

#### LITT BRANDED TRANSISTORS

_					
TTL 74 Ser Plastic 7400 7401 7402 7403 7404 7405 7406 7407 7408	0·16 0·16 0·16 0·16 0·16 0·18 0·18 0·51 0·18	TRA Type gain BC107 5 25 100 LINE	NPN. 7/8/9. pcs for pcs 3-56 AR IC	TORS Med Voltage To-18. Simil 0·35. 1·20. 3.	
7409 7410	0·18 0·16	LINE A 307	AR .	ICS	0.55
7412 7413	0·25 0·25	380		14 Pin Dii	0.90
7414	0.72	555		8 Pin Dil	0.45
7416	0.43	565		14 Pin Dil	2.00
7417	0.43	566		8 Pin Dil	1.50
7420	0.18	567		8 Pin Dii	2.00
7422	0.38	709		8/14 Pin Dil	0.35
7423 7425	0.40	741		8/14 Pin Dil	0.28
7427	0.48	748		8 Pin Dil	0.38

0-48 748 0-53 3900 0-16 CA3045 0-37 CA3046 0-35 CA3130 0-35 MC1304 0-16 MC1307P 0-76 MC1458P 0-65 MC1496L 14 Pin Dil 0.85\* 0.50 1:69 1 · 56 1 · 17 7438 7440 7441 7445 7445 7446 7451 7453 7454 7460 7472 7473 7474 7475 7476 7480 7480 7481 0.77 MC1496L 0.65 1.50 2.56 0.81 0.81 0.85 0.16 0.18 1-10 SN75324 SN75451 0.80

SN75452 TAA300 TAA310A TAA550

TAA611B12 0·18 0·32 0·26 0·30 0·32 0·47 1.25 **TBA530** CARBON FILM RESISTORS ŧ₩ ŧ₩ F24SFR E24SER

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82	0.75	TRANSISTORS
83	1.12	
84		
85		
86	0.32	@ 3 Amps.
89	2.92	G 3 Amps.
90	0.45	
91	0.68	10 pcs for £1:30

20 pcs for £3-48 50 pcs for £7-58 100 pcs for £13-00 Special Offer

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REGULATORS 0.45

120 0 40	
1 amp Plastic	
7805	1
7812	1
7815	1
7818	1
LM340-5	1
LM340-12	- 1
LM340-15	1
LM340-18	1
2.5 - 22 - 600-400 75	

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0.02

IC SOC	KETS
8 Pin	0.16
14 Pin	0.16
16 Pin	0.18
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#### TRANSISTORS. DIODES,

	1	R	ECTI	FIE	RS	
-	AC126 AC127 AC128	0·15 0·16 0·13	BD181 BD182 BD183	0·86 0·92 0·97	OA90 OA91 OC41	0·08 0·08 0·15
)	AC1281 AC141	0·18	BD232 BD233 BD237 BD238	0.60° 0.48° 0.55° 0.60°	OC44 OC45 OC70	0·15 0·12 0·10 0·10 0·10
	AC176 AC176	0.28 0.16 0.25 0.18 0.25	BD184 BDY20 BDY38 BDY60 BDY61	1·20 0·80 0·60 0·60 0·65	OC71 OC72 OC84 SC40A SC40B	0·10 0·22 0·14 0·73 0·81
)	AC187 AC188 AC188 AC188 AD140 AD142	0.90	BDY62 BF178 8F179	0·55 0·28 0·30 0·10*	SC40D SC40F SC41 A	0.98 0.65 0.65
	AD143 AD149 AD161 AD162	0·46 0·45 0·35 0·35	8F195 BF190 BF197 BF224J	0·10° 0·12° 0·12° 0·18°	SC4tF ST2 TIP29A	0.85 0.80 0.20 0.44
7	AL102 AL103 AF114 AF115	0.95 0.93 0.20 0.20	BF244 BF257 BF258 BF337	0·17° 0·30° 0·35 0·32	TIP30A TIP31A TIP32A	0·52 0·54 0·64
7	AF116 AF117 AF118	0.20	BFW60	0·17° 0·26 0·30 0·23	TIP34 TIP41 A TIP42 A IN2069 IN2070	1·05 0·68 0·72 0·14
	AF139	0·50 0·33 0·37 0·09 0·09	BFX30 BFX84 BFX85 BFX88 BFY50 BFY51	0·23 0·25 0·20 0·20	IN2070 IN4001 IN4002 IN4003 IN4004	0·16 0·04* 0·06* 0·06* 0·07*
r	BC107 BC107B BC108 BC109 BC109C	0-09 0-09 0-12	BFY52 BFY64	0·18 0·19 0·35	N4004 IN4005 IN4006 IN4007	0.08.
	BC117 BC125 BC126 BC141	0·19° 0·18° 0·20°	BFY90 BR100 BFY39	0.65 0.20 0.40	2N696 2N697	0:10° 0:14 0:12
	BC141 BC142 BC143	0·28° 0·23 0·23	BSX19 BSX20 BSX21	0·16 0·18 0·20	2N706 2N929 2N930	0·10 0·14 0·14
1	BC144 BC147	0.30	BSY95A BT106	0·12 1·00	2N1131 2N1132	0·15 0·16
	BC148 BC149 BC152	0.09° 0.09° 0.25°	BT107 BT108 BT109	1·60 1·60 1·00	2N1304 2N1305 2N1711	0·45 0·45 0·18
ğ	BC153 BC157	0·18° 0·09°	BT116 BU105	1.00 1.80°	2N2102 2N2369	0·44 0·14
	BC158 BC159 BC160	0.09° 0.09°	BU105/0 BU126 BU204	1 · 60° 1 · 60°	2N2369 A 2N2484 2N2646	0·14 0·16 0·30
	BC161 BC168B BC182	0·38 0·09* 0·11*	BU208 BY206 BY207	2·60° 0·15 0·20°	2N2905 A 2N2905 A 2N2926R	
50 50 50	BC182L BC183	0·11° 0·10°	BYX36- 300	0.12*	2N29260 2N2926Y	0.09°
50 35	BC183L BC184 BC184L	0·10* 0·11* 0·11*	900 1200	0·15* 0·18* 0·21*	2N2926G 2N3053 2N3054	0·10* 0·15 0·40
50 35 35 35	BC207B BC212	0.12*	BYX38- 300	0.50	2N3055 2N3440	0·50 0·58
35	BC212L BC213	0·11° 0·12°	600 900	0·55 0·60	2N3442 2N3525	1 · 20 0 · 50
1	BC213L BC214	0·12* 0·14*		0·65 Series	2N3570 2N3702	0·80 0·10°
1	BC214L BC237	0·14° 0·16°	Zeners BZX83 o	0·20	2N3703 2N3704	0·10* 0·10*
1	BC238 BC300	0·16° 0·34	BZX88 Series		2N3705 2N3706	0·10° 0·10°
	<b>BC301</b> <b>BC323</b>	0·32 0·60	Zeners C106A	0·11 0·40	2N3707 2N3714	0·10° 1·05
I	<b>BC327</b> <b>BC328</b>	0·18° 0·16°	C106B C106D	0.45	2N3715 2N3716	1.15
4	BC337 BC338	0·17* 0·17*	C106F CRS1/05	0.35	2N3771 2N3772	1·80 1·60
۱ ا	BCY30 BCY31	0·55 0·55	CRS1/10 CRS1/20	0.25	2N3773	2·10 0·28°
1	BCY32	0.60	CRS1/40		2N3819 2N3904	0·16°
	BCY34	0.55	CRS1/60 CRS3-05		2N3906 2N4124	0·11° 0·14
	ECY38 ECY39	0·5; 1·15	CRS3-10 CRS3-20	0.50	2N4290 2N4348	0·12 1·20
1	BCY70 BCY71	0·12 0·18	CRS3-40 CRS3-60	0.60	2N4870 2N4871	0·35° 0·35°
	BCY72 BD615	0·12 0·55	MJ480 MJ481	0·80 1·05	2N4919 2N4920	0·40° 0·50°
	BD131	0.36	MJ490	0.90	2N4922	0.58*
١	BD132 BD135	0·40 0·36	MJ491 MJE340	1·15 0·40°	2N4923 2N5060	0·64* 0·20*
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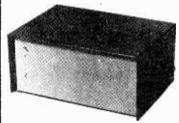
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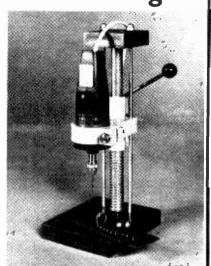
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114		F25 95	25Z5 ·80	E80CF	5 40	EF83	1 . 20	KTW61	1.75	PY801	40	ł
2D21		F28 · 70	25Z6 -80	ESOF	2.40	EF85	-35	N78	4.75	QQV03/10		Ł
3A4		36 -55	30A5 -80	E88CC	1 - 40	EF86	-40	PABC80	-40			ı
304		H6 ·50	30C15 -75	E180F	1.10			PC86	- 60	QQV03/2	5-00	ł
3Q5	-65 6		30C17 -75	E182CC	3-10	EF89	.30	PC88		OOV06/40		ı
3S4	42 6		30C18 1 00	E188CC	2 40	EF91	· 50	PC95	-60		8-00	l
			30F5 -65	EABC80	40	EF92	· 50		- 65			ı
5R4	90 6					EF97	-85	PC97	- 35	QS95/10	- 95	ı
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5Y3		N7 -65	30L15 ·70	EB91	15	EF184	- 35	PCC85	-45	UAF42	- 85	I
5 <b>Z</b> 3		Q7 -45	30L17 ·70		75	EF804	1 - 60	PCC88	- 60	UBC41	· 45	١
5Z4		R7 -60	30P12 ·70	EBC81	-40	EH90	-45	PCC89	-45	ÜBC81	- 50	l
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6AC7		SJ7 ·55	30PL1 1:00	EBF83	- 40	EL34	- 85	PCF80	- 40	UBF89	-35	ı
6AG5		SK7 -50	30PL13 1:00	EBF89	-40	EL35	3 - 20	PCF82	-40	UCC84	-85	l
6AH6	-68 65	SL7 · 50	30PL14 1-25	EC86	-80	EL37	3 20	PCF84	- 65	UCC85	-40	ł
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6AN8	-65   62	X4 -40	35W4 ·50	ECC81	- 32	EL83	-35	PCF801	-45	UCL82	-45	ı
6AQ5		X5 -40	35Z4 ·70	ECC82	.32			PCF802	-50	UCL83	-55	ı
6AR5		OC2 -65	35Z5 ·75	ECC83	.32	EL86	. 55	PCF802	- 50	UF41	- 65	ı
6AS7		0F1 65	50C5 ·70	ECC84	- 35	EL95	-65			UF42	-75	ı
6AT6		0F9 -65	50CD6 1-10	ECC85		EL360	1 - 60	PCF808	1-00		-40	ł
6AU6		0F18 -85	50L6 1-10	ECCOS	- 35	EL506	1.10	PCH200	1.00	UF80		ł
BAV6	-45 10	0LD11 ·70	85A2 ·70	IECCOD	1 · 20	EM80	- 50	PCL82	-35	UF85	-48	l
6BA6		DP13 ·75	90 AG 2-8	ECCOO	- 50	EM81	- 55	PCL83	-45	UF89	- 40	ı
6BE6		0P14 2:35	90C1 -80	ECC189	.75	EM83	- 55	PCL84	-45	ÜL41	-65	ì
6BH6		2AC6 -75	90CG 2-85	ECC807	1 · 20	EM84	- 40	PCL86	- 45	UL84	-43	ı
6BJ6		2AE8 -75	807 1.00	ECF80	- 50	EM85	1.10	PCL805/8		UM80	- 55	l
6BO7		2AT6 -50	5763 1.50	ECF82	· 50	EM87	1 - 10	PFL200	-85	UY41	-45	l
6BR7		2AU6 -50	6057 1.06		-75	EY51	- 40	PL33	-45	UY85	· 35	ı
		2AV6 -50	6060 1.00		5 - 20	EY81	-40	PL36	- 55	Ū10	- 90	۱
6BR8		2BA6 ·50	6067 1.00		-70	EY83	- 55	PL81	-45	Ū12	1 - 00	ı
6BS7		2BE6 -50	7193 -5		- 30	EY84	1 - 10	PL81A	-50	U14	1-00	ı
6BW6			CL33 1 8		- 50	EY86/7	-35	PL82	-35	U19	3-90	ı
6BW7		2BH7 ·50	CY31 6		- 50	EY88	- 50	PL83	-40	U25	.70	ı
6C4		2BY7 -80			- 40	EY91	- 50	PL84	- 45	U26	- 55	ı
BC5		2E1 3.00			-40	EZ40	-50	PL500	- 80	U191	- 50	ı
6CB6		9AQ5 60	DAF96 -34		.70	EZ40 EZ41	-50	PL504	-80	VR105/30	- 50	۱
6CD6		9G6 5·95	DF91 -3				-30	PL504	1 - 50	VR150/30	- 45	ı
6CL6		0D1 ·65	DF96 -5		- 60	EZ80			-50	W729	1-15	1
6EW6		0F2 -80	DK92 1 · 0		.70	EZ81	- 30	PY33				1
6F6	-54 2	OL1 1-10	DK96 -6		-40	GY501	- 80	PY80	- 50	X41	1 . 00	1
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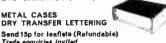
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3A/100V* 3A/400V* 3A/600V*	20p 27p	Noise D	<u> </u>	DIAC.	

$\begin{array}{cccccccccccccccccccccccccccccccccccc$	ALUM. BOXES with lid* 2½x5½x1½* 54p 4x4x1½* 54p 4x2½x1½* 70p 4x5½x1½* 70p 4x5½x1½* 54p	De-Soldering Braid* 65p Solder- sucker Tool* 475p SPEAKERS
12x5x3" 127p   16 Ω 5W	5x4x2" 86p 6x4x2" 91p 7x5x2\frac{1}{2}" 99p 8x6x3" 126p 10x7x3" 149p	2"; 2½" 65p 2·5; 3" 58p 64 Ω 2·5" 65
	10x41x3" 117p 12x5x3" 127p	7" x 4" 190p 16 Ω 5W

2030	· / X 4	203p
ER CLA	D BO	RDS*
single sided	dble/ sided	S.R.B.P. 71"x81"
115p	160p	40p
	ER CLA single sided 64p	ER CLAD BO/ single dble/ sided sided 64p 78p

FERRIC Anhydr	C C	HLO 65p	RI +	<b>DE*</b> 30p	11b p.	å	ag p.
DALO	ET	СН	RE	SIS	T .		_

DALO PEN*	ETCH	RESIS1	75p
VEROE	OARD*	Pitch	
	0.1	0.15	0.15
1	(copper	clad)	(plain)
21 x 31	″ 36p	29p	19p
1 24 x 5"	43n	39p	24p
31 x 31	" 43p	40p	_
3 x 5"		53p	34p
2 x 17		107p	73p
3 x 17		143p	95p
4 4 x 17		_	145p
Pkt of 3	6 pins		28p
Spot fac	e cutter		69p
Pin inse	ertion too	I	89p

320p

3A/100V* 18p BB105B 30p 40430 99p 32 x 17" 3A/400V* 20p BB106 45p 40669 99p 42 x 17" Pkt of 36		95p 145p 28p
3A/600V* 27p Noise Diode DIAC* Spot face 3A/1000V*30p Z5J 105p ST2 25p Pin inser	cutter	69p
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OPTO 9N5777 54p LM323K 625p	LM327N ±5 -12	260 p
130EATORS 180P	LM317K 20W 1 · 5A	200
	+1.2V to 37V	350p
TiL114 110p 75491 80p 1A -5V 220p TiL117 164p 75492 90p 1A -12V 220p	LM317H 5W 1·5A +1·2V to 37V	32 <b>0</b> p

	KNOBS* fit 1" shaft with grub scr except K2 (push fit) & K8 (for slid	
ı	K1 Black or White pointer type	9p
1	K2 Slim silvered aluminium	10p
1	K4 Black serrated. Metal top with	100
J		
1	line Indicator 33mm dlam.	22p
ı	K4A As above but 25mm dlam.	20 p
k	K5 Black fluted metal top and skirt	•
ı	calibrated 0-10, 37mm dlam.	26p
1	K6 PK2 as K5, pointer on skirt	26p
1		Zep
1	K7 Black, knurled, tapered. Metal	
4	top & skirt. Calib. 0-10, 30mm	25p
-	K8 Black or silvered for slider pot	10p
1	K9 Solld Aluminium, Amplifier K	noh.
ì	Professional type, with etch	line
1	ladicates 46 5 y 40 mm dien	96.
1	Indicator 16.5 x 12mm dlam.	∠e D

K8 Black or slivered for silder pot 10
K9 Solid Aluminium, Amplifier Knol
Professional type, with etch lin
Indicator 16-5 x 12mm dlam. 26
K10 As above but tapered 18-5 x 17mi dlam. 28
K11 Aluminium, (Top Hat) Knurle
body with 18-5mm skirt. Knotche
Attractive 30

JAC	K PLUG	S	S	OCK	ETS
	creened chrome 10p 14p 19p 28p	Plast body 8p 10p 15p 18p	me	tal p	Moulded with break contacts 20p 24p
DIN 2 pin Loud	sneaker		Plugs	_	Sockets
3, 4, 5 (180	& 240°)		12p		<b>8</b> p
CO-AXIA	L (TV)		14p		10p
PHONO Metal scre		1	9p 2p		2, 3,6way 8p 10p 20p
WANDER		1	5p		15p
BANANA	4mm 2mm 1mm	1	0p 5p 5p		12p 15p 15p

PANEL N	IETERS' Full	scale
59 x 46 x 3	5mm req. 11/1 h	ole
0-50 µ A	0-100mA	50μA-0-50μA
0-100 µA	0-500mA	100µA-0-100µA
0-500µA	0-1 Amp	500µA-0-500µA
0-1mA	0-50V DC	£3.85 each
0–5mA 0–10mA	0-300V A C	EDGEWISE
		89 x 32 x 70mm
0-50mA	"'VU"	0-1mA, 0-500µA
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108 x 82·5	x 38mm reg. 60	mm panel hole
0-504A; 0	-1004A; 0-500/	4A £5-30
BILLA 1 3711	METEROWA	

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-1mA 0-50V DC £3:85 each	MES HOLDERS Chrome cover, Red or
-5mA 0-300V AC EDGEWISE	Amber, Jewelled top 50p LES OR MES Batten Holders 10p
89 X 32 X 70mm	MES BULBS 3-5V 6V 12V Sp
-50mA "VU" 0-1mA, 0-500μA Price £3⋅55 each. Price £4⋅95	NEONS Mains, Sealed with Resistor.
08 x 82·5 x 38mm req. 60mm panel hole	Sq. Top, Red or Grn. Round Top Red 24p
-50μA; 0-100μA; 0-500μA £5·30	Neon with leads, 95V AC (No resistor) 9p
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RANSFORMERS* (Mains Prim. 220-240V)	TOGGLE: 2A, 250V 1A DP 14p SPST 26p 1A DP C/O 16p DPDT 35p 4A DP (mini) 13p
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-0-9V 75mA 95p 8-0-18V 1A 275p+	SUB-MIN TOGGLE PUSH BUTTON
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-12 U-12V 15UMA 14Up	DPDT 6 tag 78p SPDT C/over 69p
-6 0-6V 280mA 150p 9-0-9V 2A 270p+ -12 0-12V 0-5A 240p+ 30-25-20-0-20-	DPDT Centre off 92p DPDT 6 Tag 89p
5 0 15V 0.5A 220-1 05 00 04 405 4	CWITCHES DIIGH BILTTON.

V 1A 245p+ 1A 275p+ V 1A 295p+	DF 4 F SI SF
1·5A 345p+ 2A 270p+ 20-0-20-	DF DF
A 497p+ 54p	MI
Min. O/P Prl. ec. 3·2 Ω 54p	R(
lin. O/P Pri. iec. 8Ω 42p	RC

	SWITCHES PUSH BUTTON:
)	Miniature Non Locking Push to Make 15p Push to Brea
	ROCKER (white): 10A 250V
•	SP changeover centre off
	ROCKER: (black) on/off 10A 250V ROCKER: Illuminated (white)
	lights when on, 3A 240V

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	Push to Make 15p Push to Break	25p
	ROCKER (white): 10A 250V	
	SP changeover centre off	28p
	ROCKER: (black) on/off 10A 250V	23p
ı	ROCKER: Illuminated (white)	
ł	lights when on, 3A 240V	52p
	ROTARY: (ADJUSTABLE STOP) 1	ole/
	2-12 way, 2p/2-6W, 3p/2-4W,4P/2-3W	37p
ı	ROTARY: Mains 250V AC , 4 Amp	42 p

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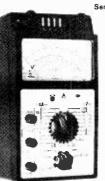


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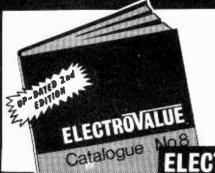
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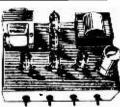
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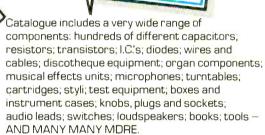
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