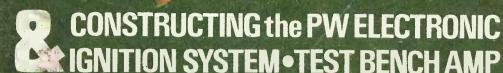
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HIGH-FIDELITY STEREO PACKAGE

Four fully wired units ready to

★ SUPER 30 AMPLIFIER (15 + 15

watt) in veneered housing

GARRARD SP25 MK III Turn-

★ GARRARD SP25 MK III Turntable on Plinth with cover
★ GOLDRING G850 Magnetic cartridge with diamond stylus
★ PAIR OF STANWAY II Speaker Units
Special Total Price Carr. £1·50

Terms: Deposit £12.75 and 9 monthly payments £9·37 (Total £97·08).

★ Super 30 Amplifier (15 + 15 watt)

★ Super 30 Amplifier (15 + 15 watt)
in veneered housing
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Turntable on Plinth as illustrated
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M Price Total £110.33

payments £10.62 (Total £110.33).



Matching as recommended for optimum performance. Package prices apply providing all individual units are purchased from any branch within 3 months. See leaflet.

★ TA12 AMPLIFIER
6.5 + 6.5 watt in veneered
housing
★ GARRARD SP25 MK III Player

★ GARRARD SP25 MK III Player
unit on Plinth
★ GOLDRING CS90 Ceramic P.U
Cartridge with diamond stylus
★ PAIR OF DORCHESTER
Loudspeaker Units

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9 monthly payments £6·25 (Total £65·25). Carr. £1·25 Trans. Plastic Cover £3·15 extra.

PACKAGE AS ABOVE but with Garrard 3000 Autochanger and Sonotone 9TA Ceramic Car- 0 1 7 tridge in lieu of SP25 and CSP25 Carr. £1.25

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'YORK' HIGH-FIDELITY 3 SPEAKER SYSTEM

★ Moderate size only 25×14×10in. ★ Response 30-20,000 c.p.s. Impedance 15 ohms

COMPLETE KIT £23 Carr. 65p

* Performance comparable with units costing considerably more.

Consists of (1) 12in. 15 watt Bass unit with cast chassis, Boll rubber cone surround for ultra low resonance, and ceramic magnet. (2) 3-way quarter section series cross-over system (3) 8 × 5in. high flux middle range speaker. (4) High efficiency tweeter. (5) Appropriate quantity acoustic damping material. (6) Handsome Teak veneered cabinet. (7) Circuit and full instructions. Terms: Dep. 24-60 and 9 monthly navuents \$9.47 (70.14 \$9.83)

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Individual Ganged Controls: Bass, Treble, Volume and Balance. Printed circuit construction employing 10 Transistors plus Diodes. Output rating I.H.F.M. Frequency range 20-20,000 c.p.s. Bass Control ± 12db. Treble Control ± 13db. Selector switch for P.U. or Tape/Radio. For loudspeaker output impedances of 3 to 15 ohms. For standard 200-250v. A.C. mains operation. Attractive Black and Silver finished metal facia plate and matching control broks.



control knobs.

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Consisting of matched 12in. 11,000 line 15 Watt 15 ohm high quality speaker, cross-over unit and tweeter. Smooth response and extended frequency range ensure surprisingly realistic reproduction. OR SENIOR 15 WATT INCLUDING HF126 15,000 LINE SPEAKER

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Heavy construction. Latest high efficiency ceramic magnets. Treated Cone surround. "D" indicates Tweeter Cone providing extended frequency range up to 15,000 c.p.s. Impedance 3 or 8-15 ohms. PLEASE STATE CHOICE. Exceptional performance at low cost.

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at low cost. HF808T 8" 10W £2-88 HF102D 10" 10W £3-40 HF120 12" 15W £4-25 HF120D 12" 15W 24-75 HF126 12" 15W 25-50 HF126D 12" 15W 25-90

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R.S.C. TA12 MKIII 6.5+6.5 WATT STEREO AMPLIFIER

FULLY TRANSISTORISED, SOLID STATE CONSTRUCTION HIGH FIDELITY OUTPUT OF 8-5 WATTS PER CHANNEL

BIGH FIDELITY OUTPUT OF 8-5 WATTS PER CHANNEL

Designed for optimum performance with any crystal
or ceramic Gram. P.U. cartridge, Radio tuner, Tape
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on each channel * Separate Bass and Treble controls
* Slide Switch for mono use * Speaker Output
3-15 ohms * For 200-250v. A.C. mains * Frequency
Response 20-20,000 c.p.s. —2dB * Harmonic Distortion 0-3% at 1,000 c.p.s. Hum and Noise —70dB * Sensitivities (1) 50mV (2) 400mV
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LES Size 16 × 11 × 9in. Pressurised. Gives pleasing results with 45.35 with any 8in. Hi-Fi speaker. Size 22 × 15 × 9in forted.

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TYPE BM1. An all-dry battery eliminator. Size $5\frac{1}{4} \times 4\frac{1}{4} \times 2in$. approx. Completely replaces batteries supplying 1.5v and 90v, to battery radio where A.C. mains 200/250v. 50c/s is available. COMPLETE KIT 43.25 ASSEMBLED READY 43.75 WITH DIAGRAM 43.75

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200-250v. AC mains operated. Frequency Response 30-20,000 c.p.s.

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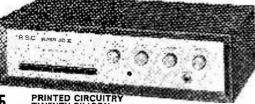
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11 ohms.

12 ohms.

13 ohms.

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15 ohms.

16 ohms.

17 ohms.

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1



CN.240/2 Miniature soldering iron 15 watt 240 volts, fitted with nickel plated 3/32" bit and packed in transparent display box. Also available for 220 volts. Price £1.70

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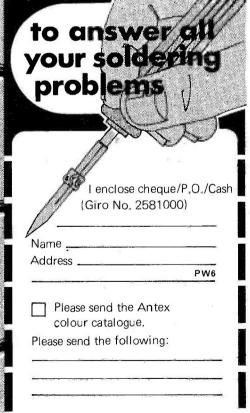
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This kit contains a 15 watt 240 volts soldering iron fitted with a 3/16" bit, nickel plated spare bits of 5/32" and 3/32", a reel of solder, Heat Sink, 1 amp fuse and booklet "How to Solder"



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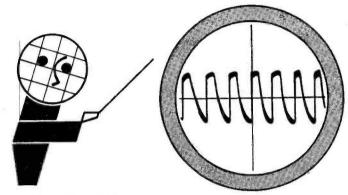
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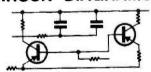
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Code Nos. mentioned above are given as a guide to the type of device in the Pak. The devices themselves are normally unmarked.

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Q33	3	1 × 2N1132	0.50
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BP02 = SN7402	0.15 0.14	0.12	BP91 = SN7491AN			0.78
BP03 = SN7403	0.15 0.14	0.12	BP92=8N7492	0.67		0.58
BP04 = SN7404	0.15 0.14	0.12	BP93=BN7493	0.67		0.58
BP05 = SN7405	0.15 0.14	0.12	BP94-SN7494	0.77		0.68
BP07 = 8N7407	0.18 0.17	0.16	BP95 -8N7495	0.77		0.68
BP08-8N7408	0.18 0.17	0.16	BP96 = SN7496	0.77		0.68
BP09 = 8N7409	0.18 0.17	0.16	BP100 = 8N74100	1.75		1.55
BP10 = SN7410	0.15 0.14	0.12	BP104=SN74104	0.97		0.88
BP13=8N7413	0.29 0.26	0.24	BP105 = 8N74105	0.97		0.88
BP16-SN7416	0.43 0.40	0.38	BP107 = SN74107	0.40		0.36
BP17 = 8N7417	0.43 0.40	0.38	BP110=SN74110	0.55	0.53	0.50
BP20 = SN7420	0.15 0.14	0.12	BP111=8N74111	1.25	1.15	1.00
BP30 = SN7430	0.15 0.14	0.12	BP115=8N74118	1.00	0.95	0.90
BP40 - 8N7440	0.15 0.14	0.12	BP119-8N74119	1.35	1.25	1.10
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BP42 - SN7442	0.67 0.64	0.58	BP141-8N74141	0.67		0.58
BP43 = SN7443	1.95 1.85	1.75	BP145=8N74145	1.50		
BP44=SN7444	1.95 1.85	1.75	BP150=SN74150	1.80		1 60
BP45 - SN7445	1.95 1.85	1.75	BP151=SN74151	1.00		0.80
BP46=SN7446	0.97 0.94	0.88	BP153=8N74153	1.20		0.95
BP47 = SN7447	0.97 0.94	0.88	BP154=SN74154	1.80		1.60
BP48 = 8N7448	0.97 0.94	0.88	BP155 - SN74155	1.40		1.20
BP50 = SN7450	0.15 0.14	0.12	BP156-8N74156	1.40		1.20
BP51 = 8N7451	0.15 0.14	0.12	BP160 = SN74160	1.80		1.60
BP53-SN7453	0.15 0.14	0.12	BP161=SN74161	1.80		
BP54 = SN7454	0.15 0.14	0.12	BP164 = SN74164	2.00		1.80
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BP60-BN7460	0.15 0.14	0.12	BP181 - SN74181	2.7		
BP70 - SN7470	0.29 0.26	0.24	BP182 = 8N74182	0.9		
BP72 = 8N7472	0.29 0.26	0.24	BP190 = 8N74190	3.5		
BP73-8N7473	0.37 0.35	0.32	BP191 = 8N74191	3.5		
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BP75 = SN7475	0.47 0.45	0.42	BP193 = 8N74193	2.10		
BP76 = 8N7476	0.43 0.40	0.38	BP195 = SN74195	1.10		
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	BP936	13p	12p	11p
1	BP944	13p	12p	11p
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$UIC936 = 12 \times \mu A 936$		$UIC9094 = 5 \times \mu A 9094$. 50p	
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$\begin{array}{llllllllllllllllllllllllllllllllllll$	UIC04 = 12 x 7404N	50n	$UIC51 = 12 \times 7451N$	50p	$UIC90 = 5 \times 7490N$	50p
$\begin{array}{llllllllllllllllllllllllllllllllllll$			$UIC53 = 12 \times 7453N$	50p	$UIC91 = 5 \times 7491N$	50p
$\begin{array}{llllllllllllllllllllllllllllllllllll$			UIC54-12 x 7454N	50p	$UIC92 = 5 \times 7492N$	50p
UIC20=12 x7420N 50p UUC70=8 x7470N 50p UIC94=5 x7490N 50p UIC95=5 x7494N 50p UIC95=5 x7498N 50p UIC95=5			UIC60 = 12 x 7460 N	50p	UIC93 = 5 x 7493N	50p
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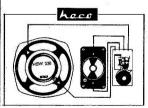
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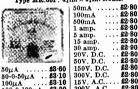


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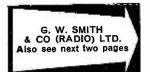
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2G374 20p 2N3569 25p 28501 2G381 22p 2N3570 125p 28502	32p BC134 12p BFX84 25p NKT242 20p 85p BC135 12p BFX85 30p NKT243 62p	CA3014 124p FJH231 25p SN7451 20p CA3018 84n FJH241 25p SN7453 20p	1U4 30p 30FL1 75p EY86 40p 1U5 60p 30FL12 120p EY87 42p
2N388A 49p 2N3572 97p 2S503 2N404 20p 2N3605 27p 3N83	27p BC136 15p BFX86 25p NKT244 17p 40p BC137 15p BFX87 25p NKT245 20p	CA3018A FJH251 25p SN7454 20p 110p FJJ101 50p SN7460 20p	2D21 35p 30FL14 95p EZ40 55p 3Q4 50p 30L15 85p EZ41 50p
2N696 15p 2N3606 27p 3N128 2N697 15p 2N3607 22p 3N140	70p BC138 20p BFX88 20p NKT261 20p 77p BC140 35p BFX89 62p NKT262 30p	CA3019 84p FJJ111 50p SN7472 30p CA3020 128p FJJ121 60p SN7473 40p	384 35p 30L17 80p EZ80 27p 3V4 48p 30P12 80p EZ81 29p
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2N718A 30p 2N3691 15p 40309 2N726 30p 2N3692 18p 40310 2N727 30p 2N3693 15p 40311	32p BC157 15p BFY30 40p NKT401 87p 45p BC158 11p BFY41 50p NKT402 90p 35p BC159 12p BFY43 62p MKT403 75p	CA3028B FJY101 25p SN7493 87p 105p IC10 250p SN7495 87p CA3029 87p IC12 250p SN7496 87p	6AG7 40p 35Z5 50p PC88 60p 6AK5 35p 50B5 50p PC97 45p 6AK6 60p 50C5 50p PC900 48p
2N727 30p 2N3693 15p 40311 2N914 17p 2N3694 18p 40312 2N916 17p 2N3702 10p 40314	47p BC160 35p BFY50 20p NKT404 55p 27n BC167 11n BFY51 20n NKT405 75n	CA3029A L900 40p SN74107 52p	6AL5 20p 80 55p PCC84 40p 6AM6 30p 85A2 50p PCC85 40p
2N918 30p 2N3703 10p 40315 2N929 22p 2N3704 11p 40316	87p BC168B 10p BFY52 20p NKT406 62p 47p BC168C 11p BFY53 15p NKT451 62p	CA3030 137p L923 40p 135p CA3035 122p LM380 122p SN74154	6AQ5 38p 807 50p PCC88 55p 6AS6 40p 1625 50p PCC89 50p
2N930 20p 2N3705 10p 40317 2N987 40p 2N3706 9p 40319	87p BC169B 11p BFY56A 57p NKT452 62p 55p BC169C 12p BFY76 42p NKT453 47p 47p BC170 12p BFY77 57p NKT713 20p	CA3036 72p MC724P 60p 200p CA3039 82p MC780P 247p SN74160 CA3041 109p MC788P 146p 180p	6AT6 35p 5763 70p PCC189 55p 6AU6 25p 6146 160 PCF80 30p 6AV6 30p AZ31 55p PCF82 34p
2N1090 22p 2N3707 11p 40320 2N1091 22p 2N3708 7p 40323 2N1131 25p 2N3709 9p 40324	32p BC171 15p BFY90 65p NKT717 42p 47p BC172 15p BSX19 17p NKT734 27p	CA3042 109p MC790P 124p 8N74161 CA3043 137p MC792P 66p 280p	6AV6 30p AZ31 55p PCF82 34p 6BA6 25p CY31 35p PCF84 60p 6BE6 30p DAF91 30p PCF86 60p
2N1132 25p 2N3710 9p 40326 2N1302 17p 2N3711 12p 40329	37p BC175 22p BSX20 15p NKT726 35p 80p BC177 20p BSX21 20p NKT773 25p	CA3044 120p MC799P 66p SN74164 CA3045 122p MC1303L	6BH6 75p DAF96 45p PCF800 80p 6BJ6 50p DF91 22p PCF801 50p
2N1303 17p 2N3713 187p 40344 2N1304 22n 2N3714 200n 40347	27p BC178 20p BSX26 45p NKT781 30p 87n BC179 20p BSX27 47p OC16 50p	CA3046 81p 200p SN74165 CA3047 137p MC1304P 225p CA3048 204p 225p SN74192	6BQ7A 40p DF96 45p PCF802 50p 6BR7 90p DK91 40p PCF805 80p 6BR8 70p DK92 55p PCF806 70p
2N1305 22p 2N3715 123p 40348 2N1306 25p 2N3716 130p 40360 2N1307 25p 2N3773 240p 40361	40p BC182L 10p BSX60 82p OC20 75p 40p BC183 9p BSX61 62p OC22 50p	CA3049 160p MC1305P 175p CA3050 185p 386p SN74193	6BW6 85p DK96 50p PCF808 75p 6BW7 80p DL92 35p PCL82 35p
2N1308 25p 2N3791 206p 40362 2N1309 25p 2N3819 34p 40370	50p BC183L 9p BSX76 15p OC23 60p 32p BC184 11p BSX77 20p OC24 60p	CA3051 134p MC838P 175p CA3052 165p 549p TAA241	6BZ6 40p DL94 48p PCL83 65p 6C4 33p DL96 45p PCL84 45p
2N1507 17p 2N3820 55p 40406 2N1613 20p 2N3823 50p 40407 2N1631 35p 2N3854 27p 40408	57p BC184L 11p BSX78 25p OC25 40p 40p BC186 25p BSY24 15p OC26 25p 52p BC187 27p BSY25 15p OC28 60p	CA3053 469 MC1435P 162p CA3054 109p 345p TAA242 CA3055 240p MC1552G 425p	6CD6 125p DM70 40p PCL85 40p 6CL6 50p DY86 32p PCL86 45p 6CW4 65p DY87 33p PFL200 65p
2N1632 30p 2N3854A 27p 40409 2N1637 30p 2N3855 27p 40410	55p BC212L 12p BSY26 17p OC29 60p 62p BC213L 12p BSY27 15p OC35 50p	CA3059 165p 461p TAA243 150p CA3064 120p MC1709CG TAA263 75p	6F1 62p E88CC 100p PL36 55p 6F6G 35p E180F 100p PL81 50p
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2N1701 163p 2N3856A 35p 40468A 2N1711 24p 2N3858 25p 40528 2N1889 32p 2N3858A 30p 40600	35p BCY30 27p BSY32 25p OC42 25p 72p BCY31 30p BSY36 25p OC44 15p 57p BCY32 50p BSY37 25p OC45 12p	FCH131 50p PA222 280p TAA320 72p FCH141 105p PA230 140p TAA350 175p	6F15 65p EB91 20p PL84 40p 6F18 50p EBC41 55p PL500 75p 6F23 85p EBC81 30p PL504 80p
2N1893 37p 2N3859 27p 40603 2N2147 72p 2N3859A 32p AC107	50p BCY33 25p BSY38 20p OC46 15p 30p BCY34 30p BSY39 22p OC70 15p	FCH151 105p PA234 92p TAA435 147p FCH171 105p PA237 210p TAA521 132p	6H6 17p EBF80 40p PY32 55p 6J4 50p EBF83 40p PY33 68p
2N2160 57p 2N3860 30p AC126 2N2193 40p 2N3866 150p AC127	20p BCY38 40p BSY43 50p OC71 12p 24p BCY39 60p BSY51 82p OC72 12p	FCH181 105p PA246 150p TAA522 360p FCH191 105p PA424 235p TAA530 495p FCH201 180p PA264 190p TAA811 445p	6J5 25p EBF89 32p PY80 40p 6J5GT 30p EBL21 60p PY81 30p
2N2193 A 42p 2N3877 40p AC128 2N2194 27p 2N3877 A 40p AC151 2N2194 A 30p 2N3900 37p AC152	18p BCY41 15p BSY53 87p OC74 30p	FCH211 180p PA265 200p TAB101 97p FCH221 180p SN7400 20p TAD100 150p	6J6 20p EC86 60p PY82 35p 6J7 45p EC88 60p PY83 38p 6K8G 40p ECC40 65p PY88 40p
2N2217 25p 2N3900A 40p AC154 2N2218 20p 2N3901 97p AC176	22p BCY43 15p BSY56 90p OC76 22p 20p BCY54 32p BSY79 45p OC77 30p	FCH231 150p SN7401 20p TAD110 150p FCJ101 180p SN7402 20p SL403D 15 p	6L6GT 45p ECC84 30p PY800 40p 6LD20 50p ECC85 40p PY801 50p
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2N2369A 15p 2N4062 12p ACY28 2N2410 42p 2N4244 47p ACY39	17p BCZ10 27p GET120 25p OC140 32p 47p BCZ11 40p GET873 12p OC170 25p	FCY101 102p SN7413 30p UA730C 160p FJH101 25p SN7420 20p UA741C 80p	68Q7 40p ECH81 30p U301 40p 6U4 65p ECH83 45p U801 £1.80 6V6G 25p ECL80 45p UABC80 40p
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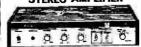
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H29	20	OA47 gold bonded diodes	50p
		coded MCS2	
N		INMARKED UNTESTED PAC	
N 866		INMARKED UNTESTED PAC Germanium Diodes Min. glass type	
1.9	EW U	MMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, PNP, Sil. and Germ.	KS
B66 B83	EW U	MMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, PNP. Sil. and Germ. Silicon Diodes DO-7 glass equiv. to 04200, 04202	KS 50p
B66 B83 B84 B86	150 200 100 50	NMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, 5il. and Germ. Silicon Diodes DO-7 glass equiv. to OA200, OA202 Sil. Diodes sub. min. IN914 and IN916 types	50p 50p 50p 50p
B66 B83	EW U 150 200 100	MMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, PNP, Sil. and Germ. Silicon Diodes DO-7 glass equiv. to OA200, OA202 Sil. Diodes sub. min. IN314 and IN916 types Sil. Trans. NPN, PNP equiv. to OC200/1 2N706A, B\$795A, etc.	50p 50p 50p
B66 B83 B84 B86	150 200 100 50	NMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, 5il. and Germ. Silicon Diodes DO-7 glass equiv. to OA200, OA202 Sil. Diodes sub. min. IN914 and IN916 types	50p 50p 50p 50p
B66 B83 B84 B86 B88	150 200 100 50 50	MMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, PNP, Sil. and Germ. Silicon Diodes DO-7 glass equiv. to OA200, OA202 Sil. Diodes sub. min. IN314 and IN916 types Sil. Trans. NPN, PNP equiv. to OC200/1 2N706A, B\$795A, etc.	50p 50p 50p 50p 50p 50p
B66 B83 B84 B86 B88	150 200 100 50 50	NMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, PNP, Sil. and Germ. Silicon Diodes DO-7 glass equiv. to OA200, OA202 Sil. Diodes sub. min. IN914 and IN916 types Sil. Trans. NPN, PNP equiv. to OC200/1 2N706A, B3Y95A, etc. Germanium Transistors PNP, AF and RF	50p 50p 50p 50p 50p 50p 50p
B66 B83 B84 B86 B88 B1	150 200 100 50 50 50 40	NMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, PNP, Sil. and Germ. Silicon Diodes DO-7 glass equiv. to OA200, OA202 Sil. Diodes sub. min. IN914 and IN916 types Sil. Trans. NPN, PNP equiv. to OC200/1 2N706A, BSY95A, etc. Germanium Transistors PNP, AF and RF 250mW. Zener Diodes DO-7 Min. Glass Type Mixed volts, 1½ watt Zeners	50p 50p 50p 50p 50p 50p 50p
B66 B83 B84 B86 B88 B1 H6	150 200 100 50 50 50 40 25	MMARKED UNTESTED PAC Germanium Diodes Min. glass type Trans. manufacturers' rejects all types NPN, PNP, Sil. and Germ. Silicon Diodes DO-7 glass equiv. to OA200, OA202 Sil. Diodes sub. min. IN914 and IN916 types Sil. Trans. NPN, PNP equiv. to OC200/1 2N706A, B\$795A, etc. Germanium Transistors PNP, AF and RF 250mW. Zener Diodes DO-7 Min. Glass Type Mixed volts, 1½ watt Zeners Top hat type 3 amp. Silicon Stud Rectifiers.	50p 50p 50p 50p 50p 50p 50p 50p
B66 B83 B84 B86 B88 B1 H6 H10	150 200 100 50 50 40 25 20	NMARKED UNTESTED PAC Germanium Diodes Min. glass type Tram. manufacturers' rejects stil types NPN, PNP, Sil. and Germ. Silicon Diodes DO-7 glass equil-v. to 0A200, OA202 Sil. Diodes sub. min. IN914 and IN916 types Sil. Trans. NPN, PNP sil. Trans. NPN, PNP squirv. to OC200/1 2N706A, BSY9SA, etc. Germanium Transistors PNP, Af and RF 250mW. Zener Diodes DO-7 Min. Glass Type Mixed volts, 1½ watt Zeners Top hat type 3 amp. Silicon Stud Rectifiers, mixed volts.	50p 50p 50p 50p 50p 50p 50p 50p 50p

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OC44	0.13	25034	0 25
OC45	0.10	2N3055	0.50
OC71	0-10		0.30
OC72	0-10	Diodes	
OC81	0 · 13	AAY42	0.10
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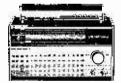
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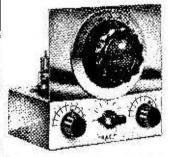
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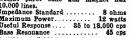
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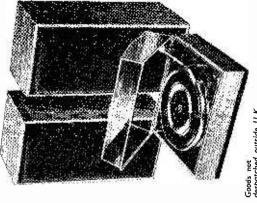
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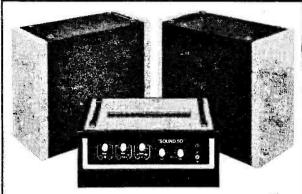
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the quality and power you need.
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Speaker Impedance: 3, 8 or 15 ohms. Bass Control Range: ± 13dB at 60Hz. Treble Control Range: ± 12dB at 10KHz. Inputs: 4 inputs at 5mV into 470K. Each pair of inputs controlled by separate volume control. 2 inputs at 200mV into 470K.
To protect the output valves, the incorporated fail safe circuit will enable the amplifier to be used at half power.

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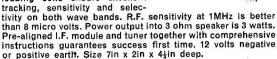
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10 TRANSISTORS. 9 TUNABLE WAVEBANDS, MW1, MW2, LW, SW1, SW2, SW3, TRAWLER BAND, VHF AND LOCAL STATIONS AND AIRCRAFT BAND

Bullt in Ferrite Rod Aerial for MW/LW. Retractable, chrome 7 section Telescopic Aerial, for peak short wave and VHF listening. Push Pull output using 500mw Transistors. Car Aerial and Tape Record Sockets. Switched Earpiece Socket complete with Earpiece. 10 Transistors plus 3 Diodes. 8° × 21° Speaker. Air Spaced ganged Tuning Condenser with VHF section. Volume on/off, Wave Change and Tone Control. Attractive Case in black with silver blocking. Size 9° × 7" × 4". Easy to follow instructions and diagrams. Parts price list and easy build plans 30p (FREE with parts).

Total building cost

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P. P. & Ins. 50p

(Overseas P. & P. £1)



ROAMER @ @ EIGHT Mk I



7 Tunable Wavebands: MW1, MW2, LW, SW1, SW2, SW3 and Trawler Band. Built in Ferrite Rod Aerial for MW and LW. Retractable chrome plated Telescopic serial for Short Wavef. Push pull output using 600mW transistors. Oar aerial and Tape record sockets. Selectivity switch. Switched carpiece socket complete with earpiece. 8 transistors plus 3 diodes. 8° × 21° Speaker. Air spaced ganged tuning condenser. Volume/on/off, tuning, wave change and tone controls. Attractive case in rich chestnut shade with gold blocking. Size 9 × 7 × 4 in. approx. Easy to follow instructions and diagrams. Parts Price List and Easy Build Plans 25p (FREE with parts).

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ROAMER SEVEN MK IV

7 Tunable Wave-bands: MW1, MW2, LW, 8W1, 8W2, SW3 and Trawler Band. Extra Medium waveband provides easier tuning of Radio Luxembourg, etc. Built in ferrite rod aerial for MW and LW. Retractable 4 section 24in. chrome plated telescopic aerial for SW. Socket for Car Aerial. Powerfol push-pull output. 7 translators and 2 diodes, including Micro-Alloy R.F. Translators. 8' × 2½' speaker. Air spaced ganged tuning condenser. Volume/on/off, tuning and wave enange controls. Attractive case with earrying handle. Size 9 × 7 × 4in. approx. Easy to follow instructions and diagrams. Parts price list and easy build plans 150 [FREE with parts). Earpiece with plug and switched socket for private listening, 300 extra.

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ROAMER





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POCKET FIVE

3 Tnnable Wavebands: MW, LW, Trawler Band with extended M.W. band for easier tuning

oband for easier tuning of Luxembourg, etc.
7 stages—5 transistors and 2 diodes,
supersensitive ferrite rod aerial, fine
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plug and awitched socket for private listening 30p

Total building costs £2.23 P.P. & (Overseas P. & P. 63p)

TRANSONA FIVE

5 TRANSISTORS AND 2 DIODES

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8 TRANSISTORS and 3 DIODES



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and Cassette Recorder PHILIPS 2204 cassette, battery/main: PHILIPS 2202 cassette PHILIPS 3302 cassette PHILIPS 4307 4-track PHILIPS 4308 De luxe 4-track PHILIPS 44074-track stereo recorder PHILIPS 44074-track stereo recorder PHILIPS 44076-track stereo recorder PHILIPS 4500 4-track stereo recorder	20 35	15.95
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TANDBERG 6041X 4-track stereo	188 - 00	145 - 25
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The Amtron range of products will shortly be available in the U.K. and a sole agent is required to handle distribution. These are just four examples from their extensive range of equipment.

TV SWEEP GENERATOR

Variable frequency signal generator for T.V. alignment; both frequency and amplitude can be modulated.

Output voltage 100mV - Frequency range: 34 to 50 MHz - Attenuator: Continuous variation - Amplitude modulation: at 1 kHz depth 30%. Can be operated on 120, 160, 220 and 240V A.C. Power Supplies.

UK 450

UK 470



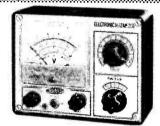
CRYSTAL CALIBRATOR MARKER GENERATOR

When a sweep signal is applied to a circuit the curves obtained on an oscilloscope are easier to analyse if they have the precise frequency reference. The UK740 marker generator provides an answer to this problem.

Ratings and characteristics; Radio frequency output voltage; 100mV Output frequency: 27-5 to 47 MHz fundamental 55 to 94 MHz second harmonic: 84 to 140 MHz third harmonic: 140 to 345 MHz fifth harmonic.

Attenuator: continuous variation - Amplitude modulation: internal 1000 Hz with possibility of cutting out the external one by UK 495 - Crystal calibrator - Output frequency: 5.5 MHz - Output

voltage at 5-5 MHz: 100mV - Power: 9V battery - Transistors used: 2 x AF106-AC128.



ELECTRONIC VOLTMETER

Using FET transistors which give better stability than the conventional valve instrument. Can be used with a 9V battery. Ratings and Characteristics - Continuous voltage range: from 20mV to 300V D.C. Bandwidth 20 Hz to 1 MHz

UK 475



BAR GENERATOR

The UK 495 can be used for setting up TV sets without the help of the broadcast test pattern.

Vertical bars: variable from 8 to 16; duration: $0.5~\mu s$ approx. - Horizontal bars: variable from 7 to 13; duration 200 μs approx. - Line synchronism: repetition rate 64 μs (15625 Hz); duration: 5 μs approx. Frame synchronism: repetition rate: 20 ms (50 Hz), duration: 600 μs approx. - Power: 9V D.C.

UK 495

Applications are invited from well established companies able to offer distribution facilities throughout the U.K.

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WIRELESS

VOL 48 NO 2

Issue 784

JUNE 1972

All the Winners!

BY a unanimous decision, the panel of judges, comprising the Editor and editorial staff of P.W., selected as the winner of the Project Autumn competition the following article:

1. Digital Frequency Counter/Timer, by J. Thornton-Lawrence (September-December 1971).

Congratulations are therefore due to author John Thornton-Lawrence for the design, construction and preparation of his most successful project since he started writing for us. Apart from the opinion of the judges, the Digital Frequency Counter/Timer has enjoyed a great success with readers, as confirmed by correspondence. John will soon be travelling to London to officially receive the Designer's Trophy and we hope that he has cleared a prominent space on the sideboard to display the cup!

The selection of runners-up was more difficult. There were twelve nominations to be considered and placed in order of merit. In the event, these were the articles selected:

- 2. Linear Ohmmeter, J. N. Watt (Sept. 71)
- 3. Quality Hi-fi System, C. R. Bradley (Feb.-April 72)
- 4. Gate Dip Oscillator, B. Wood (February 72)
- 5. Cube Radio, R. F. Graham (March 72)
- 6. Modular Audio Mixing System, F. C. Judd (Oct.-Dec. 71)

While congratulating the runners-up, we must also offer our sympathy to the several authors who very nearly squeezed into the final list—some by a very narrow margin.

As mentioned last month, and detailed elsewhere in this issue, our next writer's competition is restricted to those who have never written for P.W. before. This should provide a keen incentive to readers who have hitherto felt that the opposition might be too strong!

We will, of course, continue to publish articles by our established authors, but they will not be eligible for the competition itself. In this way, if we achieve nothing else, it will be seen that the pages of P.W. are open to anyone producing the right kind of material. A study of the winning articles in the last competition will be helpful in assessing chances and will also show that a long and complex article is not necessarily the route to a prize. And for new writers who need a little guidance on preparing articles for publication, some handy notes are available from this office on request. Now it is up to you!

W. N. STEVENS-Editor.

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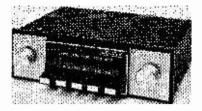
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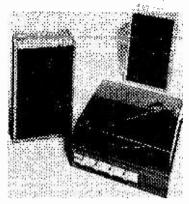
NEWS... NEWS... NEWS...

Push-button wonder



It's a "wonder" because it's a push-button car radio in kit form retailing for only £7! Step-by-step assembly instructions are provided with the kit, and the manufac-turers, Radio & TV Components Limited, maintain that anyone with a little experience in wielding a soldering iron can complete the unit in an evening—or if you go to bed very early, two evenings! An integrated circuit and a printed circuit board simplify construction and cut down the number of components to be soldered. The "Tourist" car radio, as it has been called, features five push buttons which can be tuned to four preset medium wave stations. The fifth button is for use on long wave. Spun aluminium knobs are used and the tuning scale is illuminated.

Permeability tuning is employed and r.f. sensitivity is said to be better than $15\mu V$ at 1MHz. Power output into an 8Ω speaker is better than 2.5W. The i.f. module and the tuner are pre-aligned and

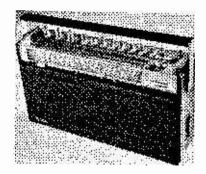


the kit is suitable for 12V positive or negative operation.

Also announced by Radio & TV Components is the £25 stereo system designated Unisound 505. This comprises pre-assembled units which can be wired together in about an hour by means of a screwdriver (no soldering iron being needed). Basically the units used in this system are the Mullard Unilex modules modified by the addition of ATES integrated circuits on the amplifier output stages to provide increased output. The turntable supplied is a Garrard 2025 TC and speakers are dual cone 13×in. elipticals in kit form, manufactured by EMI.

Further information on both of these units may be obtained from Radio & TV Components, (Acton) Limited, 21 High Street, Acton, London, W.3.

Saba radio



One receiver in the new Saba range is the Transeuropa Automatic G Radio mains/battery/car portable radio. Waveband ranges are: f.m. 87.5MHz-104MHz, SW1

6.8MHz-18.9MHz, SW2 15MHz-15.5MHz, SW3 5.9MHz-6.23MHz, SW4 2.8MHz-7.5MHz, MW1 510kHz-1220kHz, MW2 1180kHz-1630kHz, LW 150kHz-300kHz.

Twelve transistors are used and the f.m. tuner employs two transistors and vari-cap diode tuning. There are common i.f. stages for a.m. and f.m.: 460kHz 10.7MHz.

A five transistor audio amplifier provides 3.6W and 6W when operating from 12V car battery. There are 2×5 Ohm speakers—7 & 2½in.—the 2½in. speaker is switchable. Treble and base controls are provided amongst many other features. Recommended retail price is £55. U.K. Distributors are Lampitt Electronics Limited, Manchester.

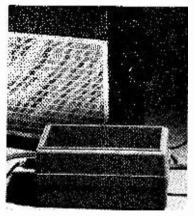
Burnt fingers

We mentioned it once before, and we're mentioning it again because it's useful stuff. Burneze, priced at 39p, is an aerosol preparation which gives instant relief to minor burns. Available through all branches of Boots and most good chemists, it should have a place in every workshop, or home first aid kit, ready for instant action the moment you grab hold of the wrong end of the soldering iron!

Hacker power

The VP408 is a fully-stabilized power unit designed to plug into the mains electricity supply and give a d.c. output adjustable from 6 to 18 volts.

Selection of the voltage required by the equipment to be operated is very simple. Set flush in the underside of the unit is a circular control which may readily be turned by a small screwdriver until the pointer on the control is in line with the voltage indicated on the scale.



The VP408 is intended primarily as an alternative to batteries as the power source for Hacker portable radios, but its wide range of stabilized voltages makes it an ideal power source for portable radios generally, some cassette tape recorders and other equipment requiring low voltage d.c. current within the capabilities of the unit. The VP408 costs £6·90p. Hacker Radio Limited, Norreys Drive, Cox Green, Maidenhead. Berks.

NEWS... NEWS... NEWS...

Record rack

The record rack shown on the front cover was kindly loaned to us by Civil Service Stores, Strand, London.

TEAC amplifier

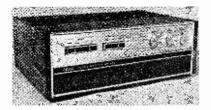
Acoustico Enterprises Ltd., inform us that they now have stocks of Teac equipment and in the picture we show the Teac Amplifier type AS-200S, retailing at £170.48. This is an all silicon transistor stereo amplifier having a rated power of 60W per channel $\pm 0.5 dB$ into a 4Ω load or 50W per channel ±0.5dB in an 8Ω load. Harmonic distortion is under 0.5%, rated output, under 0.1%, 30W output and under 0.1% 100mW. Frequency response is 20-80,000kHz +0dB -1dB and power bandwidth is 20-30,000Hz. Input impedance is $25k\Omega$ or more and sensitivity is 0.7V for rated output.

Elimination of capacitance from the output circuit plus the very stable dual p.s.u. results in an 'unprecedented' flat frequency response which is very noticeable in the low frequency ranges.

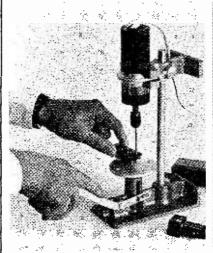
A precision 3dB step selection tone control is employed to eliminate any tonal imbalance between channels.

So that the speakers are protected from surge voltage when the unit is first switched on, Teac have included a muting circuit that prevents any current flow to the speakers until the unit has stabilized—usually 3-4 seconds after switch-on.

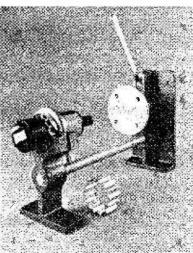
A matching and very smart looking a.m./f.m. stereo tuner designated AT-200S is also available from Acoustico for £220·41. For further information on Teac equipment contact Acoustico Enterprises Ltd., 6-8 Union Street, Kingston Upon Thames, Surrey.



Drills for printed circuit boards



Last October we gave details of two miniature electric drills, made by Expo (Drills) Ltd., which have proven very useful indeed to constructors faced with the problem of drilling a multitude of small holes in circuit boards. The Reliant, and its big brother the Titan Super, are intended to run from a 12V battery or a mains rectifier unit. Now a drillstand-cumlathe bed is available capable of accommodating either drill. The drill is attached to the stand by a fixed clamp, the work table rising up to the drill by means of a small hand lever against a spring action.



Holes are provided in the stand body to enable it to be screwed down to a bench or to a moveable wooden base. The stand can also be screwed down with the drill and main post horizontal when it may be used as a simple lathe bed for polishing, grinding or cutting. Apart from standard twist drills various cutters, burrs and saws are available as accessories. All these can be accommodated in a set of three collets with a maximum capacity of 18in. A three jaw chuck of unique design is expected to be available in the near future. Details from Expo Ltd., 62 Neal Street, London, W.C.2.

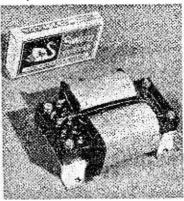
Texan transformers

Gardners Transformers Limited announce a new range of low-profile transformers which are designed to overcome the transformer problems associated with the modern tendency towards slimmer electronic equipment.

Use of a Solo Series Transformer (Gardners Type SL 20) in the "Texan" stereo-amplifier designed by Richard Mann, of Texas Instruments, enabled the designer to achieve a remarkably compact design.

Further details of the new SOLO Series (Drawing A.3869) are readily available to industrial users, while a separate Advance Technical Data Sheet (AT.23)

describing the Texan 20W transformer will gladly be forwarded upon request to Gardners Transformers Limited, Christchurch, Hampshire, BH23 3PN.





HIS is a project that really got out of hand!
Originally I had an old Garrard 4HF turntable that had seen better days and the intention was to put it in a home-made cabinet with a simple mono amplifier and hand it over to my daughter Estelle for her birthday. Her chief criterion of quality is the level obtainable, "the louder the better"!

In the end a second hand Garrard SP25 deck was fitted carrying a ceramic cartridge of unknown history. The "simple" amplifier finished up with ten transistors and at least "mid-fi" performance. The relatively small eliptical speaker in the cabinet has

become more of a monitor speaker, the amplifier output being taken to a larger external speaker better able to handle the amplifier output of 12W r.m.s.

THE PREAMPLIFIER

A pre-amplifier performs two basic functions. It must raise the usually low voltages fed into it to a level suitable for feeding the power amplifier, and it must be capable of altering the frequency response of the signals passing through it to fulfil two essential requirements. First, to equalise or com-

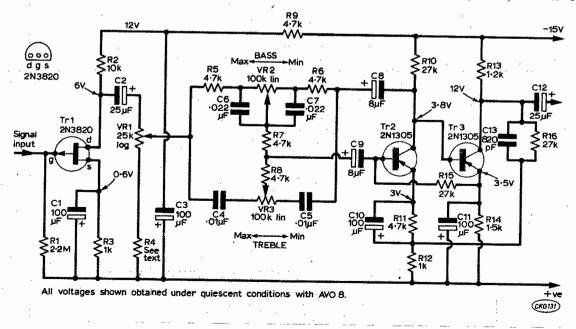


Fig. 1: The circuit of the preamplifier and tone control sections.

pensate the frequency dependent characteristics of the cartridge so that the output from the pre-amp is substantially "flat". This is achieved by providing the pre-amp with a response that is the reverse of the cartridge. Secondly, to provide a variable frequency response so that the flat response can be altered at will. Paradoxical perhaps, but a necessary requirement nevertheless.

The first requirement is a fixed frequency response to the RIAA standard. The second requirement is to provide a variable frequency response, the limits of which follow long established practice. The paradox referred to earlier is a necessary one, for the flat ideal response may result in a sound output from the loudspeaker that is aurally totally unacceptable. The variable frequency response that results from the use of correctly designed tone controls allows the user to modify the response so that a more pleasing sound can be obtained from the loudspeaker, even though a graph of the same response may well depart, perhaps greatly, from the theoretically "correct" flat line response.

Fig. 1. shows the circuit of the pre-amp used in the present design. Considering the various ways possible of loading a ceramic cartridge to provide the required correction, the simplest is by working it into a high resistance, typically $2M\Omega$, and this is the value of the gate resistor of the f.e.t. used as the input device. R2 is the drain load, the drain current of 600μ A being set by the source resistor R3, which is decoupled to a.f. by C1. The low drain current assists in maintaining a low noise level in Tr1, an essential requirement, since the following stages will amplify this noise as well as the wanted signal.

Trl feeds the volume control VRI at the earthy end of which is a low value resistor, R4, allowing a low level signal from the loudspeaker whilst a record is being played with the volume control at minimum.

This is purely a safety measure, intended to forestall the user from turning the volume right down when, for example, answering the telephone, and then forgetting all about the record player. The value of R4 is best found experimentally.

The output of the volume control feeds the tone controls which are of the well known and highly regarded Baxandall configuration. A further stage of amplification follows the tone controls before the signal is ready to be passed on to the power amplifier. The degree of amplification required is quite modest but nevertheless two transistors are allotted to this task. The configuration of Tr2 and Tr3 however, allows the overall gain (of the two) to be closely controlled, and is therefore far superior to a single transistor. Direct coupling from Tr2 collector to Tr3 base goes a long way towards eliminating low frequency phase shifts which can occasionally prove troublesome.

Base bias for Tr2 is derived from the emitter of Tr3, improving d.c. stability. Feedback from Tr3 collector to the junction of R11, R12, and C10, controls the a.c. gain and is dependent on the ratio of R16 to R12. Thus, if R16 is, for example, 20kn and R12 is $1k\Omega$, the gain will be 20 times. We can therefore control the gain within wide limits simply by varying the ratio of R12 and R16. This is a most useful facility. Since the output of ceramic cartridges can vary from around 30mV for a better class cartridge, to around a volt for the cheaper "crystal" type, it is clearly useful to be able to vary the gain somewhere in the amplifier to compensate. If this is done, then the gain can be individually "tailored", assuming the cartridge type remains unaltered, so that full output is only available with the volume control fully advanced.

A further advantage of the two transistor circuit is the ability to "tailor" the a.c. feedback such that it becomes frequency selective. This can be effected by

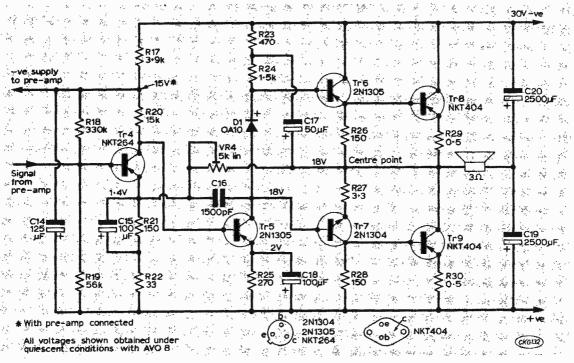


Fig. 2: The circuit of the main amplifier.

shunting R16 with a capacitor, C13. The higher the capacity of C13, the lower the frequency at which it becomes effective. C13 can be selected by purely subjective means, using a cut and try approach in conjunction with listening tests.

THE POWER AMPLIFIER

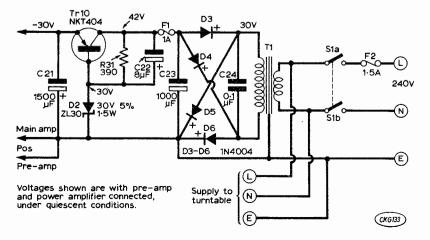
The circuit of the power amplifier, shown in Fig. 2., is entirely conventional, based upon well established and proven techniques. Several versions, both 3Ω and 15Ω output impedance, having been built and all having performed faultlessly.

Tr4, the input transistor, is a simple voltage amplifying stage with base bias provided by the potential divider R18 and R19. The collector of Tr4 is directly coupled to the base of Tr5. Feedback from the centre point of the amplifier to Tr4 emitter, via VR4, establishes the d.c. working conditions throughout the whole amplifier. Since transistors and other components are subject to manufacturing "spreads", it is necessary to include VR4 to compensate for deviations in the working conditions caused by these "spreads". VR4 is the only component used in setting up the power amplifier.

From Tr5 the signal passes to the bases of an n.p.n.-p.n.p. pair of transistors, Tr6 and Tr7. Diode D1 assists in maintaining correct bias levels. At this junction an emphatic warning must be sounded, if partially or irretrievably damaged transistors are not to be the order of the day. D1 is part of the bias network for the output transistors, and must be correctly wired into circuit. Suspect or "surplus" diodes must not be used in this application. Failure to observe these requirements may result in trouble.

Capacitor C16 introduces a degree of high frequency attenuation, in order to protect the output transistors from high level, high frequency, long duration, signals. The h.f. response of the germanium transistors used is very limited, and an excessive continuous h.f. signal can destroy them by overheating the junctions. Under what may be termed normal conditions, these problems should not arise; however, readers possessing a.f. signal generators will no doubt wish to carry out appropriate tests, and then there is a very real danger of an increased mortality rate among power transistors if due care is not exercised.

Cross-over distortion, caused as the output transistors alternately conduct and cut off, can sound most unpleasant if remedial procedures are not



★ components list

Resist	tors				
R1	2·2MΩ	R12	1kΩ	R23	470Ω
R2	10kΩ	R13		R24	1.5kΩ
R3	1kΩ	R14	1 · 5kΩ	R25	
R4	see text	R15	27kΩ		150Ω
R5	4·7kΩ	R16	27kΩ		3.3Ω
R6	4·7kΩ	R17	3.9kΩ		150Ω
R7	4 · 7kΩ	R18		R29	
R8	4-7kΩ	R19	56kΩ	R30	0·5Ω 10% WW
R9	4·7kΩ	R20	15k Ω	R31	390Ω 5% 1W
R10	27kΩ	R21	150Ω		
R11	4·7kΩ	R22	33Ω		
All ½	W 5% car	bon f	ilm exce	pt as	noted.
VR1	25kΩ tog	. 1	VR3 10	0kO lir)
VR2	100kΩ lir	· '	VR4 5k	O skel	eton pre-set
Capac			,		,
C1	100µF 15	٧W	C13	820pF	polystyrene
C2	25µF 15V	W	C14		15VW
C3	100µF 15	٧W	C15		15VW
C4	0.01µF		C16	1500r	F polystyrene
C5	0.01µF		C17		15VW
C6	0·022μF		C18		15VW
C7	0·022μF		C19	2500µ	F 25VW
C8	8μF 15VV		C20	2500μ	F 25VW
C9	8μF 15VV		C21		F 60VW
C10	100μF 15		C22		15VW
C11	100μF 15		C23		F 60VW
C12	25μF 15V	W	C24	0.1μ F	= 500 ∨ ,
Note: C21 and C23 must have an a.c. ripple rating of 1A for mono or 2A for stereo.					

Semiconductors

Tr1	2N3820	Tr5	2N1305	Tr9	NKT404
Tr2	2N1305	Tr6	2N1305	Tr10	NKT404
Tr3	2N1305	Tr7	2N1304	D1	OA10
Tr4	NKT264	Tr8	NKT404	D2	ZL30 30V 1.5W
		D3	to 6 1N	14004	

Miscellaneous

T1, mains transformer 240V/30V 750mA (mono). Fuses and holders (2). S1a-b, 2 pole rotary on-off. Knobs etc. PC board 7 x 3 in. Vereboard 0·1in. Matrix $4\frac{1}{2}$ x 3 in. Speaker 7 x 4 in. elíptical 3Ω .

taken. The standard remedy is to apply a small forward bias to the output transistors. Emitter resistors R29 and R30 perform this function in the present design.

To reduce the loading on the pre-driver stage, it is customary to increase the input resistance of the driver stage. This is effected by means of a "boot-strap" capacitor, C17 in the amplifier.

A disconcerting, though normally harmless effect found on some power amplifiers, is a loud "plop" when the amplifier is first switched on. The loudspeaker is here fed from a single output capacitor of some 2000, F from the centre point, and the

Fig. 3: The power supply circuit.

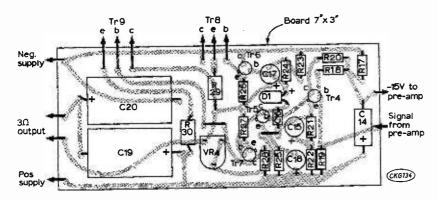
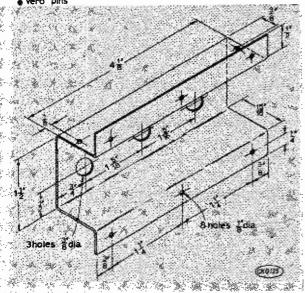


Fig. 4: The circuit board for the main amplifier viewed from the component side

"plop" is caused by the capacitor charging up to half the supply potential via the "upper" output transistor and the loudspeaker, which are effectively in series.

In the present design this effect is greatly minimised by the use of two output capacitors in series, with the loudspeaker connected from the centre point to the junction of the capacitors. We now effectively have a bridge circuit, in which Tr8 and Tr9 form one side and C19 and C20 form the other. Initial surge currents now tend to flow more symmetrically and steadier nerves should result.

A further important advantage of the symmetrical output configuration, perhaps outweighing the plop



reducing properties, is that supply ripple voltages will also flow symmetrically, resulting in a reduced hum level from the 'speaker, not that the hum level requires much reduction!

THE POWER SUPPLY

The amplifier was originally designed for a 150 output, as one of a stereo pair. The mains transformer was therefore bought with a 30V secondary to provide a rectified supply, at the quiescent current rating, of 42V. Since the present amplifier requires a supply of 30V, a series transistor in the power supply drops the surplus 12V.

The secondary of the mains transformer feeds a bridge rectifier comprising diodes D3 to D6, Fig. 3. Capacitor C24 is a bypass capacitor, intended to reduce overvoltage spikes and mains-borne noise. C23 is the reservoir capacitor. Tr10 is the series regulating transistor, the base of which is held a constant 30V by the zener diode D2.

CONSTRUCTION

The power amplifier, as already mentioned, was originally conceived as one of a stereo pair, the two amplifiers being laid out on one sheet of copper clad laminate as a mirror image pair. For this application, the sheet was split in half, shown from the component side in Fig. 4. The pre-amp was, to save time, built on a piece of 0·lin. matrix plain Veroboard, shown in Fig. 5. together with its mounting bracket.

Clearly, there is no reason why both units should not be built upon the same type of board either separately, as here, or in an integrated form. The power amplifier is not unduly critical of layout, but the straight line type of construction is undoubtedly best. The pre-amp carries and "processes" some quite low level signals, and if the signal-to-noise ratio and hum level are not to suffer, care in layout must be exercised.

The power supply is integrated to the extent of having all its components mounted on and around the mains transformer. A piece of $^{1}16$ in. aluminium 5×2 in. is bolted to one end of the mains transformer, but spaced from it by $^{1}2$ in. by means of $^{3}4$ in. 4BA bolts. In the centre of this panel is

Fig. 5: The component layout and mounting bracket for the preamplifler.

mounted Tr10 with an insulating mica washer. D2, R31 and C22 are grouped round the pins of Tr10. The ends of the aluminium panel protrude either side of the mains transformer body, and carry C21 and C23, one at each end.

The end of the transformer, remote from the power transistor heat sink, Fig. 6, carries a small panel of s.r.b.p. on to which are mounted eight tags.

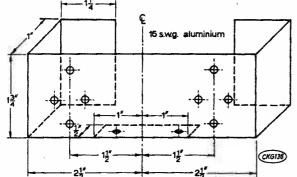


Fig. 6: The heatsink for the output transistors made from 16 s.w.g. aluminium.

Diodes D3 to D6 are soldered to these turret tags, the interconnections being on the reverse side. Since the current demands are heavy (relatively speaking) wire of a suitably heavy gauge must be used such as 14/01in.

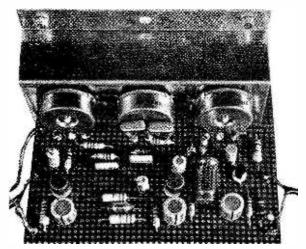
COMPONENTS

It must be borne in mind that the power amplifier and the latter two transistors of the pre-amp are directly coupled; any faulty transistors, or transistors well outside the published specifications, can lead to fault conditions developing. At best these conditions will prevent correct operation of the equipment; at worst they will cause the destruction of the power transistors. The "cure" is to buy only new transistors from a reputable supplier.

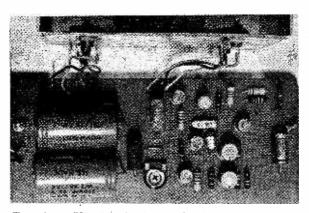
With the exception of R29, R30, and R31, all the resistors used are 5% ¹2W carbon film types since these are now freely and inexpensively available from many suppliers. The use of these resistors makes subsequent fault finding very much easier, should the need arise, since any deviation from correct operation can almost certainly be attributed to the semiconductors whose manufacturing "spreads" are so much wider than those of the resistors.

The mains transformer, as already explained, was purchased with a 30V secondary. If the expense of the simple series stabiliser is unacceptable, then the conventional power supply (transformer—rectifier—capacitor) can be used. Under these conditions, the transformer should be rated at 18 to 20V at 800mA for a single amplifier, or at 1.5A for a stereo pair. Since smoothing is now dependent upon C23 alone, it should be increased in value to $2000\mu F$ or $2500\mu F$. An anti-surge resistor of about 1Ω should be interposed between the rectifiers and C23.

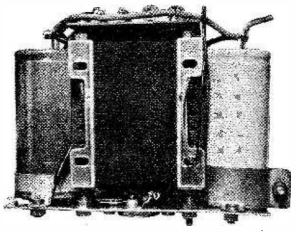
When the amplifier is delivering its full rated power, transistors Tr8, Tr9. and Tr10 get very hot under sustained drive conditions, and adequately rated heat sinks are necessary. Under normal conditions, when music is being reproduced at a reasonable level, through speakers of average efficiency, the heat generated is very much less, and smaller



A view of preamplifier.



The main amplifier. The wires to the output transistors can be seen at the top.



The power supply.

heat sinks can be used, providing an adequate air flow is ensured.

When the various units are completed, they must be thoroughly checked over; wiring errors, shorting wires, etc., can be disastrous, so time spent in careful checking is time well spent. The first item to be checked can be the power supply. Assuming a series stabiliser is used with a transformer having a 30V secondary, the voltages should agree closely with those shown in Fig. 3.

Having allowed C21 ample time to discharge completely, the power amplifier can be connected to the power supply. A fairly high wattage resistor of some 10 to 20Ω should be connected in series with the negative line to limit the maximum current that can be drawn in the event of the power amplifier having a fault condition. Power can now be applied.

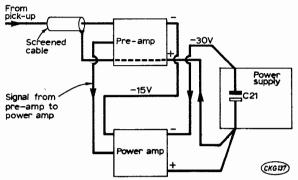


Fig. 7: The interconnection of the individual sections.

An accurate voltmeter should be connected to the centre point and VR4 adjusted to provide a centrepoint voltage of approximately 18V. The remaining voltages can also be checked and should agree with those shown in Fig. 2. If voltages differ considerably from those shown, or if the centrepoint cannot be set to 18V, then a semiconductor is the most likely cause, and the easiest course of action is to remove them all from circuit and have them checked on a reliable tester. It is no good pressing on in the hope that the fault will "go away". It won't, it will remain or get worse. If all appears to be well, the series resistor can now be removed.

The centre point voltage, under quiescent conditions, is set to exceed half the supply voltage by a small margin, so that at maximum drive the centre

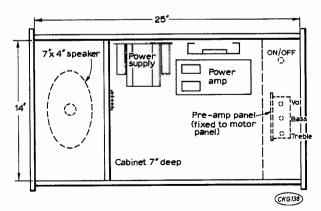


Fig. 8: The general layout employed in the prototype.

point falls to exactly half the supply voltage. Limiting, under conditions of excess drive, will therefore be symmetrical.

If an audio oscillator and oscilloscope are available, VR4 can be set even more accurately. A 1kHz

signal is injected into the base of Tr4, and the oscilloscope is connected from the centrepoint to earth. The input from the generator is increased to the point where limiting becomes visible on the oscilloscope. VR4 is then adjusted until the limiting become symmetrical, i.e. positive and negative peaks are evenly clipped. Since VR4 has an effect upon the overall gain of the amplifier, the input signal may require altering to compensate.

As explained earlier, the output transistors are liable to be damaged, or even completely destroyed if indiscriminate high power testing is indulged in. For this reason, sine wave testing at full power must not be attempted beyond 1 or 2kHz. High frequency testing can be attempted, but at greatly reduced power outputs. Even then, a series ammeter should be used to continuously monitor the current consumption. If this is done at a carefully chosen output then it will become apparent that I²R as a function of actual power output, and I²R as a function of power consumed, are two very different quantities, the power consumed being, of course, greater than the power output, the difference increasing with an increase in frequency.

When the power amplifier has been satisfactorily set up, attention can be turned to the pre-amp. Again, an audio oscillator and oscilloscope are invaluable. A voltage at 1kHz., equal to the peak output of the cartridge to be used, is fed into the input of the pre-amp and R16 is selected such that, with the volume control fully advanced, the output waveform from the power amplifier just starts to limit. This will ensure that distortion due to overloading cannot occur, even at maximum output.

CONNECTING UP

AC currents of over \$^1_2\$A flow through the output stages under maximum drive conditions and incorrect earthing can result in instability or the formation of hum loops. Some thought to interconnections must therefore be given, the circuit of Fig. 7 having been found satisfactory. The pickup lead should only be earthed at the pre-amp, the other end being left disconnected. The turntable should be earthed separately to the main earth.On the prototype, a separate mains on-off switch is used, sited well away from pre-amp to preclude any possibility of mains induced hum. This switches the mains to the player as well as to the amplifier power supply.

Although the amplifier has been described as a mono or single channel unit, reference has occasionally been made to stereo operation. This is, of course, quite possible. All that is required is a doubling up of everything, with a revised physical layout. Separate heat sinks should be used for the two power amplifiers, and if continual high power operation is envisaged, these should be of the commercially produced extruded type.

The power supply will require revision, since a single NKT404 will probably be overrun. Its maximum collector dissipation in the present application is in the region of 8W and this will be doubled for stereo. The easiest course, if a regulator is desired, is to replace the NKT404 with a 2N3055 in the positive side, and to turn the power supply circuit "up side down". The 2N3055 must be mounted on an extruded heat sink to dissipate the heat generated safely, this being mounted in a position where an adequate air flow is possible.

MODIFICATIONS TO THE TRANSISTORISED OSCILLOSCOPE

FEBRUARY-MARCH 1972

S INCE publication of the constructional details of the Transistorised Oscilloscope in the February and March editions of Practical Wireless, many readers have written in to say that they have had difficulty in obtaining the VCR 139A cathode ray tube. When the unit was first designed, stocks of this tube appeared to be adequate to meet the anticipated demand, but the response to the article has been very much greater than expected and consequently the VCR 139A has become a very rare specimen! In reply to readers' requests several alternative tubes have been tried but, to date, only two which are readily available in quantity have proved to be satisfactory—the surplus 2A P1 and the Mullard DG 7-5.

The 2A P1 can be used as a direct electrical replacement with only one circuit modification (change R12) but suffers from two mechanical disadvantages:

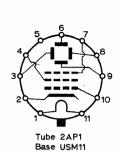
- The tube face is only 2in. diameter rather than 2³4in. of VCR 139A.
- 2. Some versions are slightly too long to be accommodated into the original chassis design. Early versions of the tube are just over 10in. long and will require extensive modification to the metalwork before they can be used. The later, and fortunately much more common version, is 7½ in. long and fits perfectly well into the cradle as designed.

The pin connections for the 2A P1 are shown in Fig. 1 together with the corresponding connections for the VCR 139A. The only electrical modification is the changing of R12 (Fig. 5 page 887 of February issue) which should now be 10Ω 5 watt.

The Mullard DG 7-5 is listed by its makers (Philips of Eindhoven) as being obsolete, but, on checking with wholesalers, it still seems to be very widely stocked throughout the country and should be available to special order from any good component retailer. This tube works very well in the revised circuit and gives, if anything, a slightly brighter and better focussed trace than the VCR 139A.

The modified display unit is shown in Fig. 2 and this should be compared with the original circuit (Fig. 5—page 887). As will be seen the main change to the original circuit is the provision of a negative supply for the tube grid. This supply is required because the heater and the cathode of the DG 7-5 are internally connected, so that instead of being able to take the grid to chassis and running the cathode at a positive potential with respect to the chassis, we are obliged to hold the cathode at chassis potential and make the grid negative. The additional components required for this modification are prefixed with the figure 4 in the revised circuit diagram.

The major points concerning construction and testing are unaffected by this modification and the few additional components are easily accommodated on the tag-strips mounted on the back panel.



DN 0669

,	2AP1	VCR139A
Heater	1,11	3,4
Cathode	2	1
Grid	10	2
Focus electrode	4	5
Anode	7	9
X plates	3,8	8,10
Y plates	9,6	7,11

Fig. 1: Pins connections of the 2AP1.

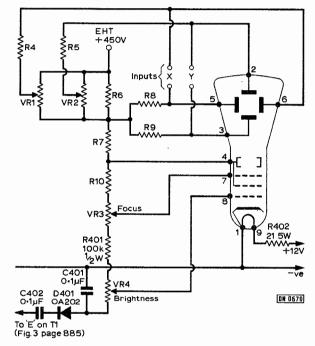


Fig. 2: Circuit changes when using the DG 7-5 tube.

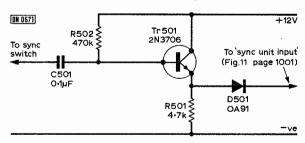
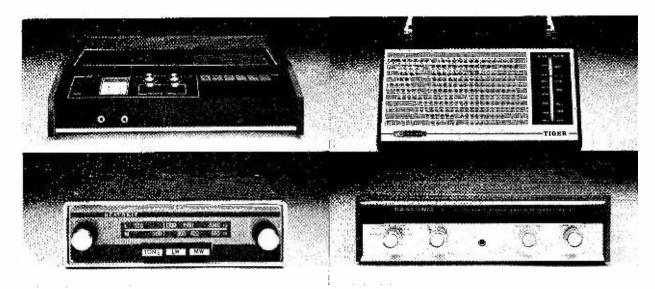


Fig. 3: Modification of the sync circuit.

Many readers have expressed a preference for a more conventional synchronisation circuit than the gated system used in the prototype. The most obvious method is to feed negative going pulses into B_2 of Tr9 (Fig. 11 page 1001) but unfortunately this seems to have a rather distressing effect on linearity. A simple and effective circuit is shown in Fig. 3—the only disadvantage being a lack of sensitivity to very fast pulses.



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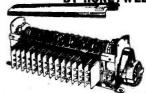
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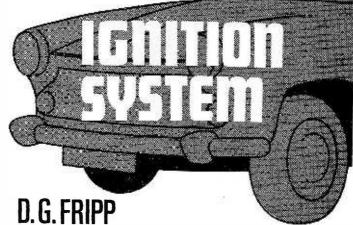
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constructing PART 1 THE PW ELECTRONIC

Last June we published a short article on the PW Electronic Ignition System little realising the tremendous interest it would arouse. Thousands of these units have been built but many readers wrote in saying "I've got all the parts, what do I do next?" Here is all the information they could possibly need, based on the experiences of a reader.

THE advantages of an electronic ignition system are now quite well known, and of the various systems available perhaps the "capacitor discharge" system is the most widely popular. This constructional article deals with such a system and follows closely the circuitry as described under this heading in the June 1971 issue in which most of the advantages were fully explained with the exception of one fact of paramount importance to the constructor; as this system uses the contact breaker as already fitted to the car, it in no way alters the timing; thus in the unlikely event of a component failure, a quick rearrangement of the connecting leads taking less than thirty seconds enables the car ignition to be returned to normal and the journey resumed.





Unlike many other systems employing specialised components which are not always readily available even from agents without some delay, this system with its inherent safety factor of easy reversion has much to commend it, as no doubt anyone who has been stranded by an ignition fault would hasten to agree.

SOME ADVANTAGES

Some not quite so obvious advantages not previously mentioned may be of interest to would be constructors, not least of which is the fact that owing to the more powerful spark delivered to the plugs by this electronic ignition unit-this may vary from 20 per cent to 40 per cent depending on the coil used-the firing of the petrol-air mixture is ensured, thus a slightly weaker mixture can often be used to enhance the economy already achieved by this form of ignition due to more complete combustion. This of course means that as up to 25 per cent more of the fuel used is burnt in the engine. where it develops more power, there is a correspondingly significant decrease in the pollution emission in the form of unburnt exhaust gases. In these days of a growing awareness of the desirability of reducing the amount of pollution, as is typified by the lowering of the limits of exhaust emission which must be complied with for the importation of cars into the U.S.A. in the near future, it would seem that the future of electronic ignition may well be assured if only from this one important aspect.

Another advantage of this type of ignition is that because of the faster rise time (as against conventional inductive systems) there is much less time for the current to leak away through any leakage paths which may exist on any of the insulated parts of the ignition components such as plug insulators, leads, distributor cap, rotor arm, and coil h.t. stack, also any conductive deposits which may have accumulated on the plug firing points. Thus plug fouling is virtually eliminated and therefore the heat range grading can be ignored. From this it can be

seen that a vehicle using this electronic ignition unit, which is driven under widely varying conditions, should not exhibit any plug trouble even if it were prone to with normal ignition, assuming plugs of reasonable condition.

Largely because of the reasons just mentioned at least one well known plug manufacturer has developed a plug specifically for use with electronic ignition. This plug has an annular gap, which has the effect of greatly increasing the mass of the outer electrode, so its erosion should be slower. It should be realised that the effect of a high energy spark is likely to cause more erosion of the electrodesespecially the outer-although this seems to be offset to some degree by using wider gaps as is advocated later in this article, and is probably nullified by the fact that this high energy spark is of much shorter duration-approximately one fifth of a conventional system—and contains nearly equal positive and negative half cycles so that metal migration of the plug firing electrodes should be very much less.

Certainly these conditions would appear to prolong spark plug life, and this is borne out by one such car fitted with new plugs when one of these ignition units was installed, as careful examination of the plugs over 5,000 miles later showed no observable deterioration of the firing electrodes.

From this it can be seen that the normal plugs as quoted by the car manufacturer can be used with confidence, provided that they are in reasonable condition and the gaps are reset.

POLARITY

The unit as shown is the one as used for positive earth supply, but the method of construction and the placing of components is identical for the negative

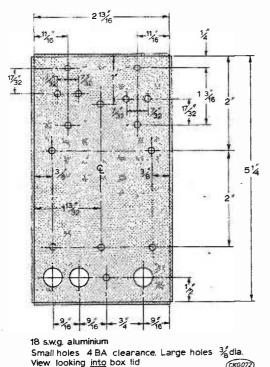


Fig. 1 Drilling details for the lid of the box on which all the components are mounted.

* components list



earth type except for the transistor types and the polarity of the associated diodes as has been explained in the original article. The actual connections from the unit to the coil and contact breaker remain the same in both models as the polarity of the pulse fed to the ignition coil is identical.

When wiring in the unit see that the pulse feed from the discharge capacitor is connected to that terminal on the coil marked positive, as this terminal should correspond to the single end of the primary winding whilst the terminal marked negative should be the end of the primary winding which is common with the earthy end of the secondary on a correctly marked coil.

CONSTRUCTION

This unit has been designed to be as compact as possible without undue cramping of the specified components, but there is little room to spare; high voltage points are adequately spaced and their positions relative to other components considered so the constructor is urged to follow the layout as closely as possible.

Having made a box and lid to the required dimensions, Fig. 1 or purchased a ready made one the lid is marked out to the dimensions given in Fig. 1 and all the holes drilled. Extra holes may be required in the box and lid if extra self tapping screws will be needed to hold the unit together when the box is mounted in the car owing to the possible inaccessibility of the one screw top and bottom provided on the ready made box.

At this stage it is prudent to ascertain the mounting position of the unit and the method to be adopted. Pan head self-tapping screws straight through the back of the box, or alternatively small pieces of aluminium angle may be affixed to the sides of the box to secure fixing points. Whichever

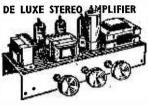
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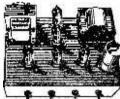
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way is used, remember that any metal protrusions in the shape of screw or rivet heads must be shallow and positioned only in close proximity to earthed parts of the circuit, and also that the unit relies on its mounting for earth return operation which keeps the number of leads to a minimum and the reversionary facility less confusing.

The three heat sink fins, having been made to the sizes given in Fig. 2, are then mounted by means of ${}^38'' \times 4BA$ screws as is also the transformer (five leads toward transistors), the two-way tag strip, the three-way centre earth leg tag strip and the discharge capacitor clip which is made from a piece of thin scrap aluminium. The eight-way tag strip is not mounted at this stage as it is more easily fitted after the tag panel.

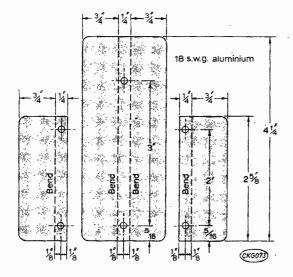


Fig. 2. These three heat sink fins are fitted to the lid of the box.

When mounting the heat sink fins and the transformer it is a sound idea to give those sides which come in contact with the chassis a smear of silicone grease to ensure thermal conductivity. Likewise the transistors when they are mounted.

To obtain a professional finish to the unit now is the time to give it a few coats of matt black paint after masking the transistor mounting positions with two small pieces of masking tape or something similar cut to the required shape using the actual transistors as templates. One of the popular aerosols is ideal for this job and the results are well worth the small effort involved, not only as regards the final appearance of the unit but this process effectively assists the even dissipation of heat besides preventing corrosion.

The paxolin strips and washers are next prepared as in Fig. 3 and assembled to form the tag panel in the manner indicated not forgetting to include a soldering tag under the nuts of connectors marked coil +, -, and CB and the earth tag of the eightway tag strip under the nut of the remaining connector marked IG/SW., the nut on that side of the transformer being removed to take the other earth tag of the eight-way tag strip which should have the adjacent tag completely removed as reference to Fig. 4 makes clear.

The two transistors of the type required for the particular polarity unit being constructed should

next be mounted (with silicone grease on mating surfaces) and also the thyristor heat sink by soldering the doubled over tag in the position indicated in Fig. 4. This heat sink may either be made from thin springy brass or copper 0.020" thick to the dimensions given in Fig. 5 or it can be formed from a commercial heat sink as supplied for OC81 type transistors as was done in the prototype. It serves as a convenient anchorage for the s.c.r. as well as providing a safety factor which, although not absolutely necessary, is an added precaution.

Wiring can now be begun and is largely achieved with the wire ends of the components. The long wire from the discharge capacitor should be well insulated with plastic sleeving and dressed as in Fig. 4. Careful checks especially on the polarity of the various diodes and the use of a heat sink is strongly recommended whilst soldering operations are in progress.

With regard to the components around the transistors, which were soldered into circuit so as to preclude any intermittent contacts which might occur with spring contacts on transistor holders (imagine an ignition fault impairing performance whilst overtaking), it may be found advantageous to solder the components together one by one so that the final connection to the appropriate transistor pin consists of only one wire end which simplifies construction, replacement of transistors should this ever be necessary, or if the unit is ever altered to the opposite polarity working.

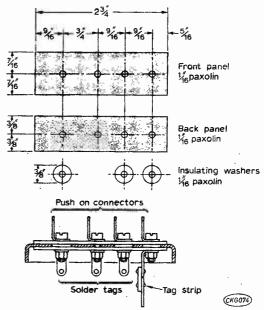


Fig. 3. Assembly details of tag panel.

The next few paragraphs may be ignored by the competent constructor, but are included for those who prefer to work from a step by step outline.

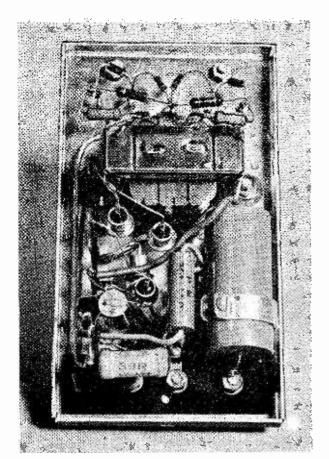
NOTE: The following instructions refer to a unit with a positive earthed vehicle. They also apply to the construction of a unit for negative earth use if the reference to polarities of D1, D2, D4, ZD1, ZD2 are reversed, D5 is not required, and of course TR1 and TR2 are of the type specified in the components list. The gate and cathode connections are also reversed for the negative earth unit.

Start by terminating the two black wires of the transformer to the third and fourth tag of the eightway tag strip (counting earth tags and removed tag from transformer towards IG/SW connector) followed by blue transformer lead and one side of R1 to right tag, and yellow transformer lead and one side of R4 to left tag of three-way strip on transistor side of transformer.

Connect one side of R2, R3 and positive side of ZD1, ZD2 to the central earth tag of the same three-way strip; join free end of R2 to free end of R1 fairly close to R1 body in a T form thus leaving the wire from R1 full length, then using this tee technique again join the positive end of D1 to the R1 wire, the end of this R1 wire is now formed into a small loop to fit on the Base pin of Tr1. By using this method there is only one connection to be made to Tr1 base pin with the advantages already mentioned.

Repeat this process with R3, R4, and D2 to Base pin of Tr2. Connect the free end of D1 (negative) to the free end of ZD1 (negative) and also green lead of T1 using same T method, again forming small loop on the ultimate wire of these junctions for connection to emitter pin of Tr1.

Repeat with D2, ZD2, red transformer lead and emitter pin of Tr2. The four BY100 which form bridge rectifier D3 are next wired in position, Fig. 4. by joining two negative wires of two BY100's together and to the earth tag on strip held by T1 bolt, one positive going to tag 3 where black T1



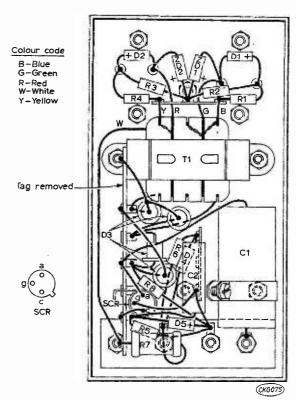


Fig. 4. Complete wiring diagram for a positive earth unit which may be compared with the photograph, below left.

lead joins. The other positive goes to tag 4 where the other black T1 lead has been soldered and from this point is connected the negative of the third BY100. Its positive wire goes to tag 5 as also does the positive of the fourth BY100 together with one side of R8. The negative of the fourth BY100 is returned to tag 3, upper half of tag. Use the lower half of tags for all previous connections except the earth one.

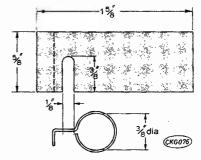


Fig. 5. Heat sink for the SCR made from springy brass or copper.

The two-way tag strip should now be mounted by means of the central heat sink fin fixing bolt and a short piece of wire soldered from the unit tag marked Coil negative to the earth tag on the two-way strip, in the form of a bus bar. Several earth joints can then be comfortably accommodated the first of which is the free end of R8.

To the lower half of tag 7 solder one end of R5 and the negative end of D5, the other end of R5 and the positive end of D5 are then earthed at the bus bar.

DISTRIBUTION PARTY OF A

PANELS

Just what you need for work bench or lab Just what you need for work bench or lab. 4 × 13 amp sockets in metal bor to take standard 13 amp fused plugs and on/ off switch with neon warning light. Supplied complete with 6 feet of flex cable. Wired up ready to work. 22-52 plus 23p P. & I. MULTI-SPEED MOTOR

MULTI-SPEED MOTOR
Six speeds are available 500, 836
and 1,100 r.p.m. and 8,000; 12,000
and 15,500 r.p.m. shaft is in. dia.
230/240V. Its speed may be
further controlled with the use of
our Thyristor controller. Very
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MAINS MOTOR Precision made — as used in record decks and tape recorders— ideal also for extractor fan, blower, heaters, etc. New and perfect. Snip at 50p. Postage 15p for first one then 5p for each one



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This comprises double-wound 230/240v mains transformer with full wave rectifier and 2000 milid* smoothing. Price 21.56, plue 200 post & packing.

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This uses the latest technique from America, a self triggering device known as the thermo tab and has enabled us to produce a really reliable dimmer at a remarkably low price—namely \$2.50 each or 10 for \$22.50.

ROCKER SWITCHES
3 new types to offer this month, all map in fixing into oblong holes. Type 1 S.P. on/off 10 amp 250V. Type 1 S.P. on/off 10 amp 250V. Type 3 D.P. on/off 10 amp 250V with neon indicator in the lever, Again Arrow 93 series. Price 25p each or 10 for \$2.25.

Type 3 D.Duble pole change over spring return, made by the French Russenberg Company. Size approx. Y. Price 15p each, 10 for £1.35.

CARD OPERATED SAFE
All electronic parts to make this \$24.50.



All electronic parts to make this 24-50,
AMPLIFIER CASE
Teak veneer on § ply, modern appearance and
design. Size—front 13' × 41' deep × 81'. Limited
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A further buy enables us to offer these at an even lower price—namely 65p each or 5 for \$2.50. Send for list of titles, We can't repeat when sold out. PRESSURE SWITCH

PRESSURE SWITCH
Made by Bailey and Macaey Ltd., Type 108R,
Adjustable up to 1510, per square Inch. (Instructions included). Set to trip at 810, per square Inch.
Changeover switch rated at 5amp 250V A.C. with
re-set button. Electrical connections in box with
conduit entry. Price \$1.50 each plus 20p post and
insurance.

20 WATT INVERTER

Smart and Brown—For van lighting or camping etc. Will light a 21t. 20 want standard fluorescent tube from a 12V car battery, current approx. 2A Very well made unit using die cast chassis. Size 11i × 2 × 1i. Price 26:50 complete with lamp holders and tube clips.

FLEX CABLE SNIP

30 core heavy circular T.R.S. waterproof flex, ideal for running down the garden to pool or shed. 1-5mm cores (5 amp) 100 yard coils \$4-25 plus carriage 75p up to 200 miles. \$1-300 miles. \$1-50 miles.



DOOR INTERCOM Know who is calling and speak to them without leaving bed. or chair Outfit comprises microphone with call push button, connectors and master intercom. Simply plugs together. Originally sold at £10. Special snip price £3.50 plus 20p postage.

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with dashboard control switch—fully extendable to 40in or fully retractable. Builable for 12V positive or negative carth. Supplied complete with fitting instructions and ready wired dashboard switch. 25 75 plus 25 post and insurance.





NEED A SPECIAL SWITCH
Double Leaf Contact.
Very alight pressure closes
both contacts. 8p each, 6y
doz. Plastic pushrod sutable for operating, 8p each,
45p doz.



NUMICATOR TUBES or digital instruments, counters, timers, cks, etc. Hi-vac XN.3. Price 98p each for £9.

24-HOUR TIME SWITCH

Made by Smiths, these are AC mains operated, NOT CLOCKWORK. Ideal for mounting on rack or shelf or can be built into box with 13A socket. Two completely adjustable time periods per 24 hours, 5A changeover contacts will switch circuit on or off during these periods. 2550 post and ins. 25p. Additional time contacts 55 pp sair.

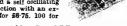


TREASURE TRACER MARK II

Complete Kit (except wooden battens) to make the metal detector similar to the circuit in Practical Wireless August issue. £2.95 plus 20p post and

MULLARD IF MODULE

This is a fully screened intermediate frequency module for amplification and detection of i.m. signals at 107MHz and a.m. signals at 470kHz. The first stage is used as an i.f. amplifier for f.m. and a self oscillating mixer for a.m. operation, in conjunction with an external oscillator coil. 75p each. 10 for \$8.75. 100 for \$82.50p. With connection dig.



2,400tt of the Best Magnetic Tape money can buy—users claim good results with Video and sound. lin wide £1:00 plus 33p post and insurance with cassette. jin wide £5p pius 25p post and insurance with cassette. jin wide £5p pius 25p post and insurance with cassette. jam and cassettes—lin 75p, jin 75p each plus 20p post and insurance with cassette. Spare spools and cassettes—lin 75p, jin 75p each plus 20p post and insurance.

ERGOTROL UNITS
These units made by the Mullard Group are for operating and controlling d.c.
Motors and equipment from A.C. mains.
Thyristors are used and these supply a variable d.c. resulting in motor speed control and operating efficiency far superior to most other methods.
The units are contained in wall mounting cabinets with front control panel on which are fuses—push buttons for on/off and the variable thyristor firing control.

4 models are available—all are brand new in makers cases:

4 models are avanance—at a continuation makers cases:

Model 2410 for up to 5 amps £17.50

Model 2411 for up to 10 amps £27.50

Model 2415 for up to 80 amps £95.00

Note: 2415 is a floor mounting 3 phase unit.





THIS MONTH'S SNIP HONEYWELL THERMOSTAT



Made by Honeywell for normal air temperatures 40°-80°F (5-25°C). This is a precision instrument with a differential which can be adjusted to better than 15°F. A mercury switch breaks on temp. rise—the switch is operated by a coiled bi-metal element and adjustable heater is incorporated for heat anticipation. Elegantly styled and encased in an ivory plastic case with clear plastic windows thermometer above and switch setting scale below-size approx 3.8° × 3.2° × 1.4° deep—can be mounted on conduit box or directly on wall. Price £1.25 each or ten for £11.25.

CENTRIFUGAL FAN



Mains operated, turbo-blower type. Pressed steel Housing contains motor and aluminium impeller. Motor is 1/10th hp giving considerable air flow but virtually no noise. Approx. dimensions 10½ wide by 12° dia. Outlet into trunking 10½° × 4½°. 24°95 + 21. THE FÜLL-FI STEREO SIX

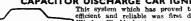


REO SIX
THE AMPLIFIER
SENSATION OF THE YEAR
You will be annazed at the
fullness of reproduction and at
the added qualities your records
or tuner will reproduce. Built
into metal chassis ready for
mounting on plint this amplifier
uses an integrated solid state
suit over the two chemps. The

uses an integrated solid state circuit with an output power of 6W R.M.S. split over the two channels. The amplifier is ideal for use with normal pick-ups and tuners, it has a double wound mains transformer and ganged volume and tone controls—also switching for Mono to Stereo, tuner or pick-up. UNREFRATABLE PRICE is \$6.50 plus 20p post and insurance. Simulated Teak cobinet ready for mounting amplifier \$1.50 (posted free when ordered with chassis). post and insurance (posted free when P.E. GEMINI Action Management

Dual purpose twin 30 watt stereo amplifier for excep-tional performance. Complete kit of parts less case \$45 or reprint of data & parts

MULLARD. AUDIO AMPLIFIER MODULE
Uses 4 transistors, and has an output of 500mW into 8 ohms speakers. Input suitable for crystal mic. or 100 of 100 of 100 of 110 of



This system which has proved to be amazingly efficient and reliable was first described in the Wireless World about a year ago. We can supply kit of parts for an improved and even more efficient version (Practical Wireless, June). Frice 25-95 plus De-lux model including printed dreut board etc. 26-95.

De-lux model including printed of RADIO STETHOSCOPE

Easiest way to fault find—traces signal from aerial to speaker—when signal stops you've found the fault. Use it on Radio. TV. amplifier, anything—complete kit comprises two special transistors and all parts including probe tube and orystal earpiece. \$28—twin stethoset instead of earpiece 75p extra—post and ins. 20p.

Where postage is not stated then orders over £5 are post free. Below £5 add 20p Semiconductors add 5p post. Over £1 post free. S.A.E. with enquiries please.

QUICK CUPPA





MAINS OPERATED SOLENOIDS



DRILL **SPEEDS**

Model TT2—small but powerful 1° pull –approx. size 1½ × 1½ × 1½ × 1½ * 60p.

Model 400/1½° pull. Size 2½ ×

2 × 1½" 75p.

Model TT10 1½" pull. Size 3 × 2½ × 2½" £1:80 plus

MAINS RELAY BARGAIN



Single pole 25p each 10 for £2:25 Double pole 35p each 10 for £3:15



DRILL CONTROLLER NEW IKW MODEL CONTROL

Electronically changes speed from approximately 10 revs. to maximum. Full power at all speeds by finger-tip control. Kit includes all control. Kit includes at parts, case, everything and full instructions. \$1:50 plus 13p post and insurance. Made up model also available. \$2:25 plus 13p post & p.

SLIDE SWITCHES



Slide Switch. 2-pole changeover panel mounting by two 6B.A. screws. Size approx. Jin × žin rated 250V lamp. 6p each. 10 for 54p, 100 for £5-10, 500 for £24. Ditto as above but for printed circuit 5p each 10 for 45p, 100 for \$4.25. Sub Miniature Side Switch. DPDT 19mm (\$\frac{2}{3}\$ in approx.) between fixing centres. 12p. each or 10 for £1.08.

LIGHT CELL Almost zero resista

Almost zero resistant in sun-light increases to 10 K Ohms in dark or dull light, epoxy resin sealed. Size approx. Iin dia. by Jin thick. Rated at 590 MW, wire ended. 43p with circuit. Also ORP12 light cell 45p.

TELESCOPIC



0-8 AMMETER 2 in square full vision for flush mounting. Moving iron instrument. Ideal for charger. Price 43p each. 10 for £3-90.



<u>J</u>oo



EXTRACTOR FAN Cleans the air at the rate of 10,000 cubic ft. per hour Suitable for kitchen, bath-Suitable for kitchen, batherooms, factories, changing rooms, etc., it's so quiet it can hardly be heard. Compact, 51' casing with 52' fan blades. Kit comprises motor, fan blades, sheet steel casing, pull switch, mains connector, and fixing brackets. \$2 plus 38p post and ins.

DIGITAL CLOCK

As featured this issue, send for parts list.

MICRO SWITCH

A changeover contacts. 9p ach. £1 doz. 15 amp Model 0p each or £1.05 doz.

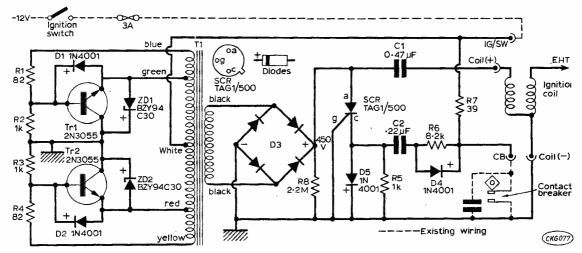


2 pole, 2 way—4 pole, 2 way—3 pole, 3 way—4 pole, 3 way—2 pole, 4 way—2 pole, 4 way—2 pole of way—1 pole, 12 way. All at 20p each £1-80 for ten, your assortment.

- REED SWITCHES Glass encase acased, switches operated by ext-gold welded contacts. We can magnet-gold offer 3 types.

J. BULL (ELECTRICAL)

(Dept. P.W.), 7 Park Street, Croydon CRO IYD Callers to: 102/3 Tamworth Road, Croydon,



Only minor amendments have been made to the original circuit. This one is for POSITIVE earth vehicles.

One end of R6, the negative end of D4, and one end of C2 are soldered to the free tag on the two-way strip; the free end of R6 and the positive end of D4 are then connected to the unit tag marked CB as is also one side of R7, the other side of R7 and the white T1 lead solder to the earth tag 8 which is unit tag marked IG/SW.

The free end of C2 connects to lower half of tag 7 on eight-way strip.

Capacitor C1 fixing clamp should now be fixed in position and C1 secured by means of the clamping bolt—not forgetting to use a small piece of aluminium as a crush guard. One end of C1 connects to unit tag marked coil positive and the other end covered with good quality plastic sleeving and dressed as in Fig. 4 and connected to tag 5 on eightway strip.

The thyristor heat sink is then fixed by bending both upper and lower portions of tag 6 (on eightway strip) in such a manner as to form an almost complete wrap around the strip, the slot in the heat sink is then pushed down centrally over this wrap and soldered in position as shown in Fig. 4.

The last component to be fitted is the thyristor

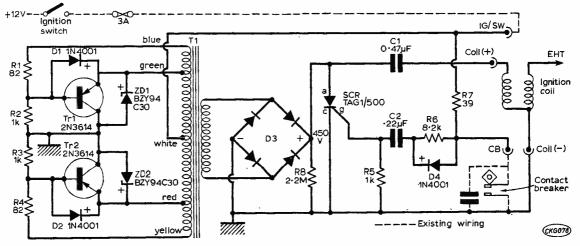
SCR and reference to Fig. 4. clearly shows how this is connected, namely anode to tag 5 cathode to tag 7 on eight-way strip whilst the gate is connected to the earth bus bar adjacent to unit tag marked Coil negative.

The TAG 1/500 has proved eminently suitable for use with most ignition coils as fitted to British cars and far superior to those s.c.r.'s obtained from other sources, in freedom from breakdown, but some foreign-made ignition coils require the TAG 1/600 to be used owing to the primary inductance generating excessive oscillatory peaks. However, this subject is too extensive to go into in this article and the foregoing remarks can be taken as a reliable guide.

TESTING

When the unit is complete it should be tested by applying the car battery to connectors IG/SW and —ve (IG/SW to live feed and —ve to that pole of the battery which is earthed on the particular car and unit) when a whistle should be heard indicating that the oscillator section is working. The voltage

This circuit is for **NEGATIVE** earth vehicles. Certain diode polarities are reversed as well as the gate and cathode connections to the SCR. Diode D5 is omitted.



across the bridge rectifier is then measured and should be approximately 450V, this reading may increase to approximately 480V when the connections to the ignition coil are made but it depends on coil characteristics.

Assuming these initial tests to be satisfactory the unit is then fixed to the car in a suitable place, all connections checked, and the unit is ready for use.

CONNECTING UP

The connections of the ignition unit to the vehicle are quite simple but are given here in full for guidance of any constructor who may be doubtful.

After mounting the unit (preferably on side of wing or bulkhead so as to keep leads from the unit to the ignition coil as short as practicable) remove the existing ignition switch to ignition coil wire—that is the thin lead which does not go to the contact breaker—and connect to unit tag marked IG/SW (see notes in text about ballast resistor if cold start coil is fitted).

Remove the thin lead from the ignition coil which connects to the contact breaker, and connect this lead to unit tag marked 'CB'.

Two short extra leads are now required and are connected one lead from unit tag marked 'coil negative' to ignition coil terminal marked 'negative,' the other lead goes from unit tag marked 'coil positive' to ignition coil terminal marked 'positive.' These connections are identical for both negative and positive earthed vehicles.

For coils marked SW and CB, this corresponds to positive and negative respectively—see circuit diagram.

Should the reversionary facility be required—due to component fault, although this should not occur if the recommended components are used—simply remove the two short leads connecting the unit tags marked 'coil positive' and 'negative' to the ignition coil positive and negative.

Remove the lead from the unit tag marked CB and connect it to the ignition coil terminal which is the same polarity as that terminal of the battery which is earthed to the chassis.

The remaining lead which is left on the unit tag marked IG/SW is removed and connected to the remaining free terminal on the ignition coil.

PLUGS

It is recommended that the sparking plug gaps be reset to around 0.050'' instead of the usual gap which is about 0.025'', as the higher resistance load thus presented enables the greater output available from the electronic ignition unit to be used to the fullest advantage.

The prototype of this unit is in use on the writer's car and has proved a worthwhile accessory.

NEXT MONTH Part 2 reviews the various types of ignition coil available today and how they can be used with the PW Electronic Ignition System plus information on using your tachometer with this unit. Finally, a discussion of the various advantages of the PW unit as obtained in actual road tests.



Practical Wireless Designer's Traphy 1972

To encourage new authors, entries for the 1972 Trophy will be restricted to readers who have not previously had an article published in PW. This leaves the field wide open for those wanting to try their hand at writing technical constructional articles. Contestants will not be in competition with well-known authors, only with other newcomers, so the cup can only be won by a new writer. It Could Be You.

TURN YOUR CONSTRUCTIONAL PROJECT INTO CASH—AND MAYBE WIN THE CUP!

- The winning entry will be chosen by a panel of judges from among articles published in issues of PW dated September 1972 to August 1973 inclusive. The Editor's decision on all matters arising will be final.
- The winner of the competition will receive and retain outright the PW Designer's Trophy 1972. Other prizes will be awarded to the best runners-up. Articles will be paid for shortly after publication.
- 3. The competition is open only to authors who have not previously had any work published in PW.
- 4. Articles submitted for the competition should conform to the general style of material published in PW and must describe the operation and construction of a piece of radio, audio or test equipment that has been designed and built by the author.
- 5. Articles should, preferably, be typed using double spacing, leaving wide margins, on one side only of each sheet. Circuit diagrams and any other drawings must be separate and numbered to agree with the text. Author's roughs must be clear enough to permit re-drawing. Components list must also be separate and laid out to the standard PW format.
- Photographs of the equipment are desirable and should be in black and white, sharp and clear. Photographs may be identified by sticking a label on the reverse instead of writing on the back of the photograph itself.
- Components used in the design must be readily available from retail sources.
- 8. Articles should be sent to the Editor, Practical Wireless, Old Fleetway House, FarrIngdon Street, London, E.C.4. Authors will be advised as soon as possible of the acceptance or rejection of their articles. Equipment, the subject of an article, must not be sent to the Editor until advised to do so.
- Employees and staff of PW are not eligible for entry to this competition.

SUBMIT YOUR ARTICLES NOW

TAKE 2®

JULIAN ANDERSON

A series of simple transistor projects, each using less than twenty components and costing less than one pound to build.

AGAZINES such as Practical Wireless require a continual supply of new ideas and new articles in order to attract readers; this is not always easy as new and original projects are rare. There is one field however which seems to have been almost ignored—that of burglar and security alarms. A few have been published but I can only recall three in the past ten years.

This is all the more extraordinary when one considers how useful (and how simple) a burglar alarm can be. Admittedly the electronics side is only the tip of the iceberg as the hard part is the fitting and wiring of the alarm circuit rather than the construction of the alarm itself. This can present problems but some guides will be given.

The circuit of the alarm itself is shown in Fig. 1. This is a simple, but reliable audio oscillator whose output is fed to a loudspeaker. The alarm circuit wire is shown as a shorting link between the base and emitter of Tr1; this cuts off the transistor. This link will, of course, be very long; it may be up to 100 yards but even so its resistance will not be more than a few ohms assuming that reasonable quality wire is used. This will have virtually no effect on the circuit and may, for our purposes, be considered as a dead short.

When this link is broken Trl will be biased by R1 and when conducting this will in turn provide bias for Tr2 causing a considerable current to be passed through the primary of the transistor output transformer T1 and so to the speaker. However the inclusion of C1 causes this current to appear as a series of pulses and this sounds like an audio note in the speaker.

When the alarm circuit is closed the only current passing will be that through R1 plus the tiny leakage currents through the transistors. This will be below $20\mu A$ in total and this sort of current can be taken from a battery almost indefinitely; it will decay of old age before running down.

Note that the output of the speaker, while being more than sufficient to scare off a burglar, is not all that high and so an efficient speaker should be used. This excludes the use of miniature types and the larger the diameter, the greater will be the output.

The construction of this circuit should present few problems but a suggested layout on a small tagboard is shown in Fig. 2.

The alarm circuit comprises a single wire running between the doors and windows to be protected. The contacts can take many forms. Microswitches can be used, most of these have changeover contacts and so can be arranged to break the circuit when a window or door is opened. A springy metal can also be used to make the contacts. All of the switches must be in series of course so that if any one of them is opened the alarm will sound.

No. 37 Burglar Alarm

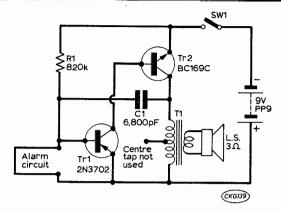


Fig. 1: The circuit of the Take 20 Burglar Alarm.

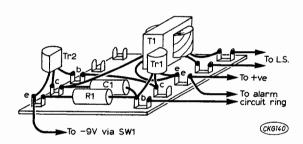


Fig. 2: A suggested component layout on tagboard.

R1	820kΩ, ½W, 5%	1p
C1	6,800pF, polystyrene	4p
Tr1	2N3702	13p
Tr2	BC169C	11p
T1 .	Transistor output transformer	
	(Eagle Type LT700 or similar)	20p
LS	3Ω loudspeaker (ex-TV type)	15p
SW1	On-off switch	5p
	Tagboard	11p
		80p
Wirel made	s are those recently advertised in Pracess and may have changed. No allowar for minimum order costs or for postaging and these should be checked carefullying.	nce is e and

In addition to acting as a burglar alarm, this arrangement will also provide a rapid check to ensure that all the necessary windows and doors are shut. It is also a fail-safe circuit; any faults in the circuit will cause the alarm to sound.

Although battery operation is perfectly satisfactory if the battery is replaced regularly, it is far better to operate this circuit from a mains power supply. In the long run this will be cheaper and more reliable.

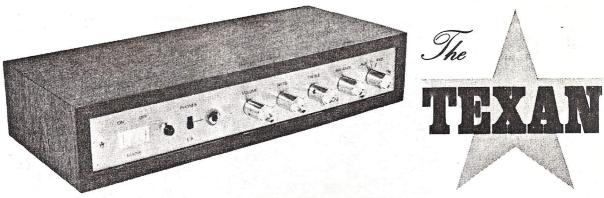


WIRELESS



This unit will greatly improve the intelligibility of weak DX signals—by filtering the a.f. rather than working at r.f. It incorporates a highly efficient adjustable noise limiter together with an audio filter which will virtually eliminate heterodynes, whatever the strength.

ALL IN THE JULY ISSUE ON SALE 2nd JUNE This simple 4-transistor crystal controlled converter permits reception of the entire 2 metre amateur band (144-146MHz) on any receiver capable of covering from 4 to 6MHz. A suitable choice of crystal frequency allows direct frequency read-out on 2 metres from the calibration on the main receiver.



20+20 WATT I.C. STEREO AMPLIFIER

PART 2

RICHARD MANN

Note: Switch section S1A in Fig. 2 (last month) should have been in the closed position. In the specification, the dynamic range should have been given as +38dB and in Fig. 7 the + and — terminals should be reversed. Under the heading "Quiescent Current Setting" on page 56, reference is made in the text to R28. This should have been R24. In the components list the BZY88C15 was specified for ZD1 and ZD2. Although this device is quite suitable for the TEXAN the Texas 1S2150A zener is now specified for ZD1/ZD2.

Equalisation

The input selector switch has three positions—for radio, magnetic pickup and an auxiliary position. The radio position gives a flat response to the input stage from <5Hz to 500 kHz and the input sensitivity is 30mV for 20watts output into 8 Ω . This is probably over sensitive for some tuners but the *Texan* has a good overload margin of some 38dB so that even

500mV from the tuner could be handled comfortably.

The pickup equalisation characteristic is shown in Fig. 13. In this position the overall gain of the amplifier is 74dB at 1kHz giving a sensitivity of $2\cdot5mV$ for full output. Also plotted is the theoretical RIAA curve which is formed from three time constants—3180 μS , 318 μS and $75\mu S$. Assuming that R6 is $270k\Omega$ (exactly!) this gives the following values for R7, C7 and C8 respectively: $21\cdot76866390k\Omega$, $3\cdot730376467nF$ and $10\cdot08777777nF$ —approximately! The values given last month in Fig. 2 are the nearest standard values and in the prototype gave a maximum error of about 1dB in the audible range. However, these theoretical values are quoted only for the theorist.

The beauty of operational amplifiers, of course, is that given the near-perfect component value you will get a near-perfect response.

This brings up the question of crystal and ceramic pickups which many people will undoubtedly wish to use. There are several ways of

tackling this:

For the cheaper cartridge giving an output of several hundred millivolts a high impedance attenuator such as shown in Fig. 14 could be used. With such a high signal there will be little lost in signal/noise performance and the $1M\Omega$ load will give a fairly flat output without further correction. This means that R7, C7 and C8 should be omitted and R6 changed to about $10k\Omega$.

Cartridges giving outputs of tens of millivolts cannot be treated this way as the signal would be attenuated too heavily since it is necessary to keep the 'earthy' resistance down to about $5k\Omega$ to avoid damping the rumble filter. The rumble filter could be omitted leaving only R2 which would be returned to pin 2 of IC1 instead of ground. It would then be bootstrapped so that the input impedance would rise to several megohms. Again R7, C7 and C8 would be omitted and R6 reduced to

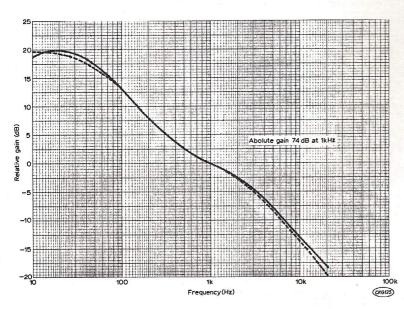
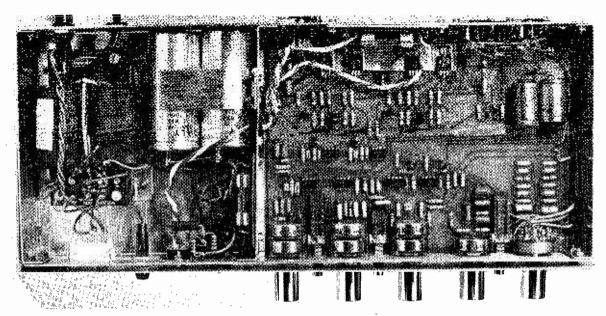


Fig. 13: Magnetic pickup characteristic (measured—solid curve, theoretical—dotted curve).



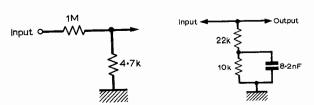
The complete Texan amplifier. The power supply section is to the left of the dividing screen.

approximately $1k\Omega$. However it seems a pity to lose the rumble filter especially as ceramic cartridges are more likely to be used with the sort of turntables which require a rumble filter most.

Therefore, the best approach is to use a low impedance loading circuit on the ceramic cartridge such as shown in Fig. 15. This will give a characteristic which approximates to the velocity characteristic of a magnetic cartridge and the output level will also be similar so that the amplifier can be used without modification to its feedback components. This circuit allows some variation of the shunt to improve the linearity of the capacitor response and some manufacturers will quote the circuit values which give the best "magnetic" characteristic from their particular cartridge. If this approach is adopted the constructor can wire the attenuator quite neatly across the pins of the DIN pickup socket.

Input Impedance

In the specification the input impedance of the amplifier is quoted as $47k\Omega$ at 1kHz. This nominal figure is modified when the rumble filter is inserted in circuit partly due to the shunting effect of R1 and partly to the series reactance of C1 and C2. This will not normally have any effect on a magnetic pickup cartridge but it may have some loading effect if a ceramic/magnetic conversion network is used so the variation of input impedance is plotted in Fig. 16.



Auxiliary

The *Texan* was originally designed to give an equalized output directly from a tape head without external amplifiers. For this reason the circuit shows R8, R9 and C9 and the printed circuit board is laid out to take these components. However, the present

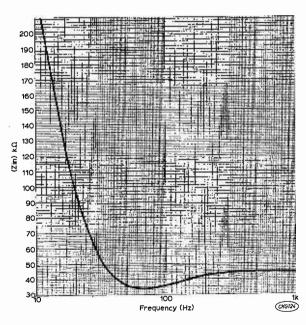


Fig. 14: (far left) Equalisation circuitry for high output crystal/ceramic pickup cartridges.

Fig. 15: (left) Equalisation circuitry for low output crystal/ceramic cartridges.

Fig. 16: (above) Variation of input impedance with frequency (rumble filter in).

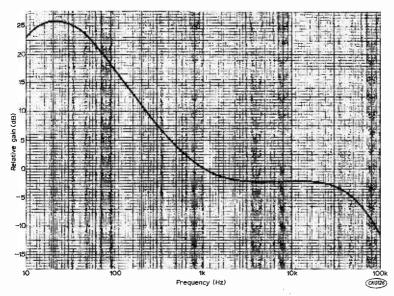


Fig. 17: Tape replay characteristic (direct from head).

day tendency is for a separate tape unit, having internal amplifiers. If it is intended to use such a unit, R8 and C9 should be omitted and R9 reduced to approximately $1\cdot 2k\Omega$ giving a flat characteristic and sensitivity similar to that of the radio position.

Nevertheless I feel that many home constructors may be interested in a tape head facility so the following table gives the appropriate values for a few of the standard replay characteristics.

TABLE 1

* * * * *	R8	R9	C9	Standard time constants
D.I.N. 13 ips D.I.N. 23 ips D.I.N. 73 ips N.A.B. 74 ips N.A.B. 33 ips	33kΩ 33kΩ 22kΩ 15kΩ 27kΩ	390kΩ 820kΩ 1 MΩ 1 MΩ		1590µS 120µS 3180µS 120µS
N.A.B. 17 ips		As fo	r 31 ips f	N.A.B.

The overall response of the amplifier when using components for DIN 354 i.p.s. is shown in Fig. 17. The sensitivity of the amplifier with this characteristic was 1mV approx. Once more the very high loop gain of the operational amplifier is valuable for producing the large amount of bass boost which is required.

Performance

A number of facts and figures have already been quoted regarding the performance of the *Texan* so that distortion is the main topic left for discussion.

Apart from the money the other good thing about writing an article is the opportunity it gives for liberating a few proverbial bees from one's bonnet. My personal "bee" is concerned with the vicious

circle of "specmanship" which sets Hi-Fi designers chasing each others' tails (or should it be "tales"). Now I am all in favour of pickups which track, noise reduction systems for tape recorders, f.m. broadcasting and electrostatic loudspeakers-on the whole I am sure that they are worth the money. But I cannot see much point in paying more and more for better and better amplifiers when they are already too good for the transducers coupled to them. When the most experienced ears can barely detect 0.1% distortion on pure tones why spend money struggling for 0.01%, especially when all transducers and recording media introduce about 2 or 3% in themselves. This is a generalisation, of course, and it is always easy to find a specific flaw in such an argument. However, I feel the basic principle is good. Namely, consider the system as a whole and don't spend a lot more money unless you are going to

hear the difference. With that said, here are a few more figures:

TABLE 2

	Power		onic o		375 - 75 345, 325
Fundamental	output	786 7	(dB)		T.H.D.
Load frequency	(watts)	2nd	3rd	Ath 4	×%:
15Ω 1kHz %	15 %	76	67	×	0.047
	10	74	80	. 86	0.023
THE WAY WE	5	74	80	80	0.024
Secretary Secretary	0.5	71	₹ 77∜	. 80 ⊴	.0-033
	0.05	59	61	62	0.164
10kHz 💉	15	1	48	80	0.488
	10	54	53	70	0.302
	5	56	57	64	0.221
	0.5	57	57	67	0.205
1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	0.05	61	70	75	0 096
8Ω 1kHz	25	63	61	69	0.114
	20		× 63		0.09
	15	66	65		0.075
1. 2. K . A. X. X. X.	5	69	75	72	0 04
	0.5		75	3. 	0 025
1. A. 15. A. W. W.	0.05	65	68	69	0.077
₩ 10kHz	25		43	63	0.779
	20	52	46	69	0:562
· 父 韩 夏 米 李	15	54	50	70	0.375
	5	× 60	56	65	0.196
	0.5	60	77	-	0.101
l e se se se se	0.05	60	المراجع والمراز	 	0:10
	30000 - 1-2		3.78 N.22		. 77.559.
4Ω 💉 1kHz 🗐	25		50	75	
1 32 W No NO NO	20	63	50		0.324
	15	65	52	****	0.257
	10		55	90	
	5	68	61	81	0 098
1. 65 米 久 海 潮	0.5	A - 40 Car .	70	71	0.066
1 10 10 10 10 10 10 10 10 10 10 10 10 10	0.05	55	58	59	0:245
B - 784 - 1782 - 785 - 785 - 785		2 2 2 1	30.00	1,70	

Distortion: Harmonic distortion was measured using a Radford Low Distortion Oscillator and a Hewlett Packard 3590A Wave Analyzer with a 3593A Sweeper.

This measurement technique is far more accurate and gives more useful information due to the subjective nature of harmonic distortion. The harmonics

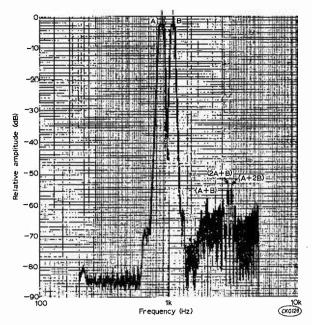


Fig. 18: Wave analysis for input frequencies of 900Hz and 1-1kHz.

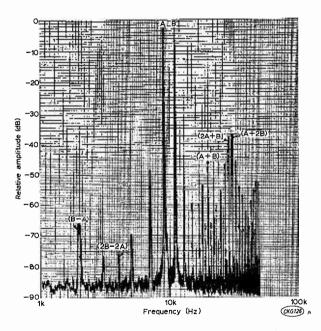


Fig. 19: Wave analysis for input frequencies of 9kHz and 11kHz

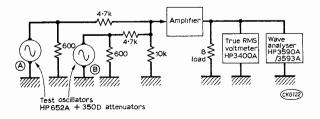


Fig. 20: Test set-up that was used for measurements of intermodulation products (I.P.).

are therefore tabulated in some detail in Table 2 along with total harmonic distortion figures. So you pays your money and you takes your choice. The harmonics are quoted in dB below the fundamental. The total harmonic figures are given as a percentage and calculated from

T.H.D. =
$$\sqrt{V_2^2 + V_3^2 + V_4^2} \cdot \dots$$

where V_2 , V_3 , V_4 etc are the percentage values of the harmonic components.

It can be seen that the percentage T.H.D. does not leap up at the levels where crossover distortion would be apparent so the amplifier has a good clean sound.

The Texan is primarily designed to work into 15Ω or 8Ω speakers but, for interest, some distortion figures are also quoted for 4Ω loads.

All the measurements were made on the complete amplifiers so they include any distortion due to the preamplifier.

Intermodulation Distortion

The intermodulation products (I.P.) in an amplifier's output result from non linearity of the transfer characteristic which causes multiplication of the components of a complex input waveform so that a spectrum of sum and difference frequencies may be produced across the entire amplifier bandwidth. Thus with only two sinusoidal inputs with frequencies A and B we may get I.P.'s at frequencies of A+B, A-B, 2A+B, 2A-B, A+2B, A-2B etc. If the spectrum is analyzed there will also be components at 2A, 2B, 3A, 3B which are due to harmonic distortion in the signal source and those harmonics produce their own I.P.s resulting in the general mish-mash shown in Figs. 18 and 19.

However, after a bit of mental arithmetic, it is fairly easy to sort out the I.P.s which really count. The total intermodulation distortion is calculated from:

I.D. =
$$\frac{\sqrt{IP_1^2 + IP_2^2 + IP_3^2}}{A + B} \times 100\%$$

where I.P., etc are the amplitudes of the intermodulation products A and B are the amplitudes of the input waveforms.

Therefore any I.P. which is 10 to 20dB below the major I.P. in level can virtually be ignored.

The method of measurement was as follows: With the apparatus shown in Fig. 20, oscillator A was temporarily disconnected and the level from oscillator B was adjusted to give 12·6 volts across the 8Ω load (ie 20 watts). The B attenuator was then set back 3dB. This procedure was repeated for oscillator A alone. The two inputs were then mixed together and the output checked on the true r.m.s. meter to ensure that the power was still 20 watts.

The input frequencies were 900Hz and 1·1kHz in one case (Fig. 18) and 9kHz and 11kHz in the other case (Fig. 19). The analyzer was set to sweep from 200Hz to 5kHz for the low frequency test and from 2kHz to 50kHz for the higher frequency test with an analyzer bandwidth of 100Hz in each case.

Fig. 18 shows that with inputs 900Hz and $1\cdot1kHz$ the predominant I.P.s occur at 2kHz (A+B), $2\cdot9kHz$ (2A+B) and $3\cdot1kHz$ (A+2B). These components give a percentage I.D. of $0\cdot19\%$ approximately.

At the higher frequencies it is easier to pick out I.P.s due to the amplifier and again it can be seen

from Fig. 19 that the dominant components are at 20kHz, 29kHz and 31kHz. The difference frequency I.P.s are also clearly seen at 2kHz (B-A), 4kHz (2B-2A) etc but they are insignificant compared with the sum products so that an I.D. figure of 1.0% is obtained.

These distortions may seem rather high but the method of measurement was rather unkind since the peak voltage for the combined waveform is $\sqrt{2}$ times greater than the peak voltage for a pure sine input due to the beating of the two waves. The peak output voltage is thus 25·2 volts giving a peak power of 80 watts instead of 40 watts.

Noise and Crosstalk

The wave analyzer which was used for the distortion measurements is also a very valuable instrument for measuring noise and crosstalk. It gives more accurate and meaningful results and since one has only to insert the graph paper and push the button it appeals to my lazy nature.

The noise versus frequency plot shown in Fig. 21 was made with the Texan switched to the radio input and the volume control turned up to nearly maximum so that the input sensitivity was exactly 30mV. The input was then grounded via 600\Omega and the wave analyzer was connected across the amplifier output.

Between 20Hz and 1kHz an analyzer bandwidth of 10Hz was used, necessitating an automatic sweep rate of 1Hz/Sec. To avoid spending six hours or so completing the plot to 25kHz, the bandwidth was increased to 100Hz after 1kHz. This allowed the sweep rate to be increased to 10Hz/Sec.

The ordinate scaling of the graph is relative to the full output voltage (12.6V) and it can be seen that between 20Hz and 1kHz the mean level

is approximately -110dB. Above 1kHz the level jumps by 10dB since noise is proportional to **/Bandwidth**

(and 20
$$\log_{10} \frac{100}{10} = 10$$
dB)

However, the absolute noise/root cycle is still the same, about $38\mu V/\sqrt{Hz}$. To get a full bandwidth signal/noise ratio we must add 33dB to the plotted level:

(i.e.
$$20 \log_{10} \frac{20 \text{kHz}}{10 \text{Hz}}$$
) giving a figure of 77dB.

The wave analyzer allows the hum components to be measured separately since peaks are obvious at 50Hz and particularly at the odd harmonics of 50Hz indicating that they originate in the power supply. Adding these components together gives a separate

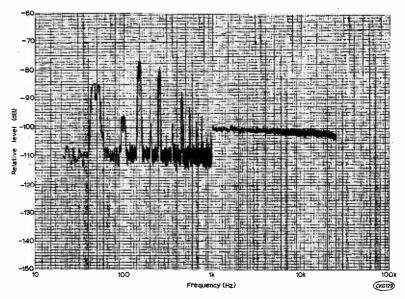


Fig. 21: Noise v. frequency.

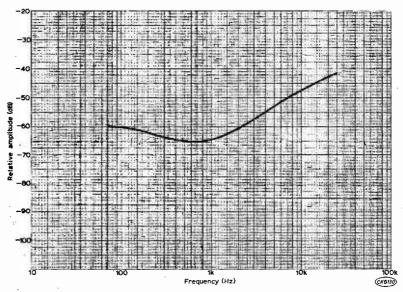


Fig. 22: Interchannel crosstalk v. frequency.

figure of 75dB for the signal/hum ratio.

More subjectively I have found that for normal listening in domestic surroundings one has to put an ear fairly close to the loudspeaker to decide if the amplifier is switched on or not and that is really the acid test.

To measure crosstalk versus frequency the b.f.o. output of the wave analyzer was used to provide a 30 mV input to one channel of the amplifier. The input of the other channel was grounded via 600Ω and the balance control was set to its mid-way position. The volume control was adjusted to give 20 watts into one load so that any extra coupling via the power supply would be included. The analyzer input was then connected to the output of the other channel. A continuous sweep was made between

60Hz and 25kHz at a bandwidth of 100Hz. The ordinates of the plot are again relative to full output voltage giving an inter-channel crosstalk figure of -65dB at 1kHz and -48dB at 10kHz. The crosstalk figures quoted in the specification were measured with an r.m.s. voltmeter at full bandwidth-hence the difference of 14dB at 1kHz. This shows the merit of using a selective voltmeter at low signal levels if other spurious signals are likely to be present. At 10kHz the crosstalk becomes the predominant signal so that the specification figure is not in error at this frequency.

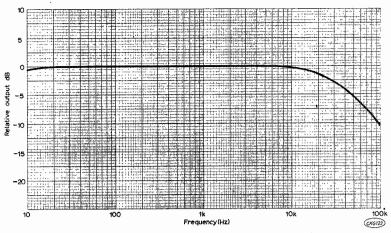


Fig. 23: Power response (20W into 8Ω resistive load).

Power response

The frequency response of the *Texan* was plotted with the selector switch in the flat radio position and with the input and gain adjusted to give a power output of 20 watts at 1kHz into an 8Ω resistive load. The response is almost identical to the low level response shown in Fig. 5 (part 1). This indicates that the gain of the amplifier is still determined by the passive feedback components in the circuit and is not effected by changes in the parameters of the transistors in the power stage. The phase advance capacitor C18 produces a smooth roll-off outside the audio band to eliminate r.f. signals from the output which could cause intermodulation problems with stereo multiplex decoders, tape oscillators and

so-forth. In response to numerous enquiries, readers are reminded that all components for the *Texan*—including drilled fibre glass p.c. board, drilled and punched metalwork, finished front panel, will be available from Henry's Radio Limited.

Kits will also contain pre-formed wire packs to facilitate assembly and complete hardwear—in fact everything will be included even down to the last nut and bolt.

A slimline version of the teak sleeve—slimmer than the type shown in the photographs will be available during July/August.

Henry's Radio Limited are the sole U.K. distributors for the "TEXAN", to the trade and retail outlets.

Larkill, Worcester.

PL5 3DU.

TO BE CONTINUED

May 7— Spalding "Tulip Time" at picnic site, Surfleet, 4 miles north of Spalding on the A16 Spalding - Boston road. Talk-in stations will be G3VPR/P on top band (1980kHz). Something for all the family. Free admittance.

May 21— Northern Mobile Rally, at Moore Grange School, Parkstone Avenue, off Ring Road, West Park, Leeds. Refreshments will be available. Further details from D. Binns, G3MGI, 80 Gipton Wood Road, Leeds 8, Yorkshire.

May 28— Chiltern Mobile Rally, organized by the Chiltern Amateur Radio Club, and held in the grounds of Sir Francis Dashwood, at West Wycombe, near High Wycombe, on the same day as an annual steam rally. Talk-in on 160m and 2m. Further details from: P. Perkins, G3OUV, Loakes House, Loakes Park, High Wycombe, Bucks. High Wycombe (0494) 21612.

May 28— Hull & District Amateur Radio Society held in grounds of the East Riding College of Education, Bishops Burton, on the A.1079 York to Beverley. Further information from L. D. Colley, G3AGX, Micasa, Ferry Road.

MOBILE RALLY DIARY

Wawne, Hull, Yorkshire.

June 11— Third Elvaston Castle, Elvaston Castle Countryside Park, Nr. Derby.

June 18— Anglian Mobile Rally, at the Suffolk Show Ground, Ipswich. Further details from D. W. Thomas, G3ZLN, The Old Peoples Home, 9 Burlington Road, Ipswich, Suffolk.

June 25— Bristol City & County RSGB Group, at Longleat, Warminster, Wilts.

June 25— West of England Mobile Rally, at Longleat, near Warminster, Wiltshire. Information from D. Iles, G3COP, 23 Dryleaze Road. Stapleton. Bristol.

July 2— South Shields & District Amateur Radio Club.

July 9— Cornish Mobile Rally, organized by the Cornish Radio Amateur Club, will be held at the Truro Rugby Football Ground. Talk-in stations will be operational

on 1.875kHz a.m. and 2m a.m.

July 16— Upton-on-Severn Mobile
Rally organised by the Worcester

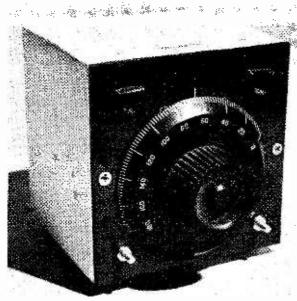
& District Amateur Radio Club.
Further information from B. A.
Jones, G8ASO, 12 Woodside Road,

August 6— Woburn Abbey Rally.
August 13— Torbay Amateur
Radio Society Mobile Rally at
Newton Abbot Rugby Ground.
August 13— Annual Derby

Mobile Rally at Rykneld Schools. Details from T. Darn, G3FGY, 1, Sandham Lane, Ripley, Derby. August 20— Saltash & District Amateur Radio Club Rally at Saltash Grammar School, with side-shows etc. Ample free parking on site. Details from: 1 Aldridge, G4AJU, 302 St. Peter's Road, Manadon, Plymouth, Devon,

August 26-27— Stratford-on-Avon Radio Club Mobile Rally at the National Agricultural Centre, Kenilworth, Warwickshire. Hq. of the 'Royal Agricultural Society of England. Further details M. J. W. Webb, G300Q, 14 Townsend Road, Tiddington, Stratford-on-Avon, Warwickshire. Or ring Stratford-on-Avon 5973.

MW/IF



WOBBULATOR

A.J.BIRKINSHAW

WOBBULATOR is a signal generating oscillator which is frequency modulated by the sawtooth timebase sweep of an oscilloscope. The band of frequency swept is designed to cover the pass band of tuned radio frequency coupled circuits such as the intermediate frequency transformers of a superhet receiver.

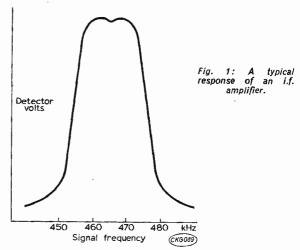
THE REQUIREMENT

The response curve of an i.f. amplifier may be plotted by gradually shifting the frequency of a calibrated signal generator of constant amplitude applied to the amplifier input and recording the detector output voltage relative to input frequency as in Fig. 1.

We could use a pen recorder to trace output voltage on a moving paper chart if there were a mechanical linkage from chart drive to the tuning control of the oscillator with linear rotation relative to frequency.

One would plot many graphs before deciding which give the best or required performance and as preset controls on i.f. transformers are not designed to give an indication of electromechanical value and in some instances a certain amount of hysteresis between electrical and mechanical value may exist, the process of resetting may become somewhat tedious.

Most of us have at some time or other taken the easy way out and tuned for maximum aural response which is reasonably effective under the circumstances because we do not have the manufacturers resources in the way of special equipment



designed to set-up a particular model.

However, because radio receivers depend largely on their i.f. response curves for the quality of reproduction, both bandwidth and amplitude have to be considered. We are in deeper trouble if we have just substituted a transformer of higher Q where maximum amplitude may lead to instability. So we require to see the behaviour of the circuit during adjustment.

We can display the band-pass response on an oscilloscope screen during alignment operations if we automatically sweep the signal in synchronism with the 'scope's time base using the detector output voltage for vertical deflection of the trace, Fig. 2.

Sweep may be obtained if the time base voltage controls the reactance of the oscillatory circuit of the signal generator to swing its resonant frequency a known amount equally above and below a mean value, the mean being the dial frequency setting of the generator.

Frequency modulated signal generators which, with the aid of a cathode ray oscilloscope, were designed for the visual examination of band-pass response curves and the alignment and testing of radio receivers have used various methods to obtain the required sweep as technology advanced. First the motor driven condenser, then the Miller reactance valve and nowadays we have the much

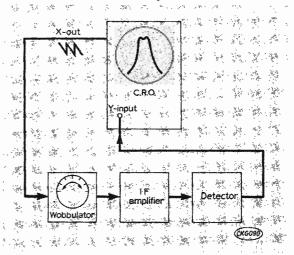


Fig. 2: The arrangement used for displaying the curve shown in Fig. 1.

simpler semiconductor diode whose reverse bias capacitance varies with voltage.

The author is indebted to D. Bollen who described, in the January 1970 issue of *Practical Wireless*, how the common silicon power rectifier shows useful varactor properties. The author has found by experiment that we may dispense with the bias battery for the oscillator application.

For amateur receiver projects the wobbulator to be described will repay its simple cost even if the oscilloscope required as an accessory has to be borrowed. Its output is 100 to 370mV peak to peak over the tuning range of 370kHz to 1.2MHz which covers normal intermediate and medium-wave frequencies.

The minimum requirements of the oscilloscope are moderate, a vertical sensitivity of 0.1 to 1V per centimeter and a time base speed range covering 10mS to 1mS per centimeter being all that is required.

THE CIRCUIT

A single OC44 germanium PNP transistor is used in a circuit configuration popularly employed as a common base oscillator found as part of self-oscillating mixers in medium-wave superhet transistor radios. The emitter resistor is smaller in value than usually employed to allow for greater r.f. output.

The diode requires a negative reverse bias which is not apparent by first examination of the circuit diagram shown in Fig. 3. The bias is provided by rectification of oscillatory power developed across R4 and R5.

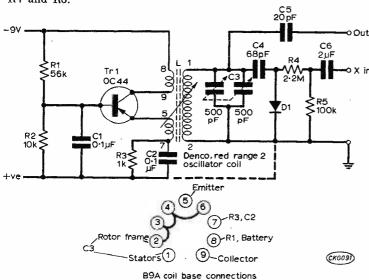


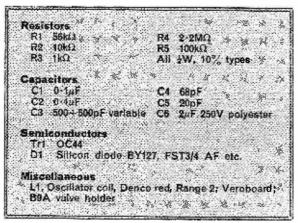
Fig. 3: The circuit of the m.w./i.f. wobbulator.

The X input in a sawtooth waveform is applied via C6 to preserve the waveshape. Sawtooth amplitude should be in the region of 20V peak to peak which is required to swing the diode capacitance and thereby frequency modulate the oscillator.

A lead via a 20pF capacitor connects to the signal input of the receiver being examined. A screened lead from the detector load connects to the Y input of the oscilloscope in use (screening is essential to avoid instability).

With the Denco coil specified, a twin gang 500pf tuning capacitor connected in parallel covers the

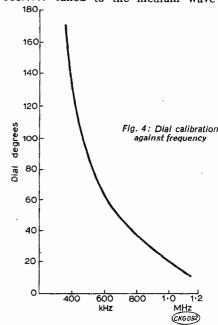
* components list



range depicted in Fig. 4. A twin 365pF unit may be substituted with an acceptable shift of frequency coverage.

CALIBRATION

The dial can be an engraved knob marked every two degrees from 0 to 180°, an alternative is to use a small protractor and pointer knob. With the case removed, connect the battery and place near a broadcast receiver tuned to the medium wave-

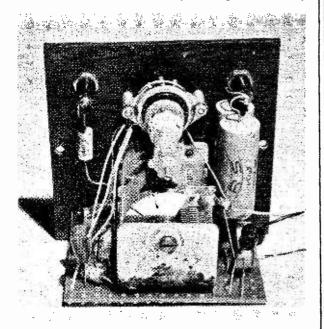


band. Select a programme of known frequency and rotate the wobbulator tuning control to obtain a beat whistle. Check with the calibration graph shown in Fig. 4 and adjust the core of the coil to obtain a similar calibration.

Construction should present no problems and the layout adopted by the author may be seen from the photographs. The Denco coil is best mounted on a B9A valveholder, this will avoid the necessity of soldering directly to the pins.

The simple wobbulator project can be put to further use by the addition of an audio modulating

oscillator operating at about 440Hz. We do not of course require modulation when using the wobbulator but there are many occasions where a modulated oscillator with reasonably pure sine wave characteristics is an asset, having successfully



An internal view of the prototype.

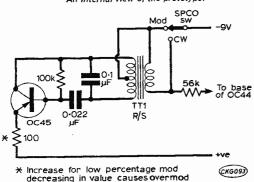
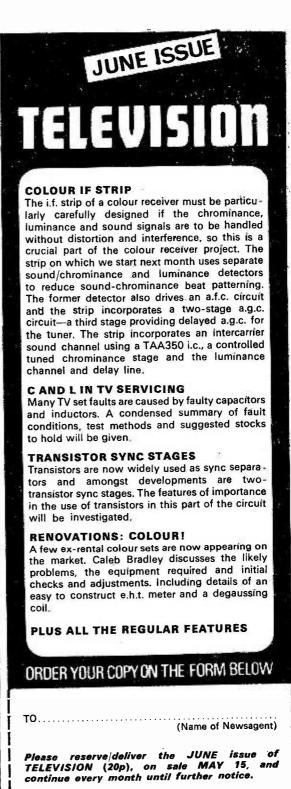


Fig. 5: A simple modulator circuit which may be added.

aligned our receiver i.f. stages we may prefer to adjust the signal r.f. stages and also to note the aural response through the amplifiers driving the loudspeaker. A tunable modulated oscillator enables us to do this, conversion is a simple matter for if we do not apply a sawtooth input, the r.f. oscillator provides c.w. output which may be base modulated by a single transistor oscillator as shown in Fig. 5.

The additional stage may be constructed on a small piece of Veroboard, this and a single pole changeover switch may easily be accommodated in the space at the rear of the cabinet.

Modulation depth may be adjusted by altering the value of the 100Ω emitter resistor but the value given is a reasonable compromise to allow sufficient tolerance in the performance of the component parts. Reducing the value gives overmodulation and poor waveform, increasing the value decreases modulation amplitude but provides a sine wave without distortion.



NAME

LETTERS The Editor does not necessarily endorse the views expressed by correspondents

On the bottle

Like your four correspondents of March '72 "Letters" I, too, was bottle-fed on valves. However, they should try to recapture the spirit of adventure and voyage into the New World of electronics. More than 50 years ago valves were also unreliable (and costly) remember? but we persevered until better days came along. Now transistors and other solid state devices have opened the door very widely indeed. Transistors unreliable? Not any more when properly used and treated. A transistor radio designed and built in 1958 by the writer is still working 100 per cent without any breakdowns. Blame the faults on designers and manufacturers of commercially made equipment built to a price-profit formula: to the lack of a new kind of servicing expertise which many persons cannot be bothered to acquire. The craze for making equipment for ordinary purposes smaller and yet smaller, added to the profit regardless, one must expect shoddy goods. No, Valvers, the day of the bottles-good as they were and blessed their memory-is gone forever. Space travel, satellite communications. etc, would not have been possible without solid state devices-nor would there have been any moon walk on television. Greatest boon of all is possibly the contributions these devices have made to Medical electronics.—A. V. Nash. (London, S.W.12).

N.Z. prices

I have just read a letter in your magazine, from K. B. Moore, dated December, 1971, and I feel I must point out that the prices for the items listed are as follows: BC109 transistor—\$0.84c (42np) not \$1.85c. BC169C transistor-\$1.10c (55np) not \$2.25c. $5\mu F$ 12V capacitor—·14c (7np) not \cdot 20c. $1 \cdot 5M$ resistor— \cdot 05c (2^{1}_{2} np) not 10c. If Mr. Moore was actually charged the prices he mentions, I would suggest that next time he buys components, he shops around first.

As you will see from the corrected list, semiconductors are about 4 to 5 times their cost in

the U.K., but passive components are very similar in price. The major reason for the difference is that the N.Z. Government places a rather large import duty and sales tax on all semi-conductors.—A. R. Millar, (Auckland 10, New Zealand).

Fight Back

As a human being, I am open to susceptance, and it is with great reluctance that I resist a tirade against "bottles," induced by the first four letters in your March issue. However, as an unbiased electronics enthusiast it would be wrong to omit to mention that valves have their uses. Indeed, only a fool would suggest using transistors in the output stages of a high power radio transmitter, for example. But semiconductor devices also have a place. Who, for instance, would even consider building a computer of valves, or even of discrete components?

In his letter in the March issue Mr. Martin gives no evidence against transistors, nor any in support of valves; it would be interesting to know the reasons for his electronic reactionism! By comparison, Mr Freeby gives difficulty of servicing as his reason for preferring valves. He mentions "transistor radios which almost fall to pieces when you try to service them," but this is the fault of their physical construction, not the transistors therein. However, he does conclude "or blow up half a dozen transistors when searching for one faulty one", and this 'fault' can be cured by practise on the serviceman's part.

"Valves are best for starting people off on electronics," writes Mr. Watton, but here I must beg differ, for the following reasons:

- (i) Transistors are very cheapthose used by beginners that is, which can be obtained for less than 1p each (the transistors, not the beginners!)
- (ii) There is no risk of electric shock or of burns in beginner's transistors circuits.
- (iii) The electrical fragility of transistors encourages care on the part of the constructor, surely

an important part of electronics.

Finally, Mr. Wode's only argument seems to be the average size of loudspeakers in transistor radio sets gviing poor tone. I am sure we all agree that large loudspeakers are essential for good quality sound, but where do transistors come in? Admittedly. transistors are electrically fragile, and a transistor class B amplifier needs careful design, for it to be successful, but the fact that all the audio amplifiers I know of on the market today are transistor amplifiers, prove that, in this field at least, transistors are supreme!—R. D. Broome, (Warwickshire).

U.S.A. — 1940

Further to your Leader in the March issue, your younger readers may be interested to know that prior to 1940, reception of U.S.A. stations in the S.W. Bands was so good that the London evening newspapers used to print the radio programmes of Boston, New York, Schenectady etc. in addition to B.B.C .-G. Snewin, (London, E.17).

Quality speakers

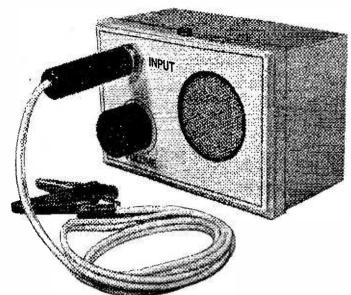
British beaches may soon have no sand judging by the number of people who seem to be building the Quality Hi-Fi speakers! But here's a useful bit of information which may help readers.

The SI-1020A hybrid amplifier specified for the system has been discontinued but may be replaced by the SI-1025A which is for all practical purposes identical in performance and mounting. The plinth specification is unchanged. Amplifers available from Photain Controls Ltd., Randalls Road, Leatherhead. Tel. 2776.—Caleb Bradley, (Essex).

Company policy

I couldn't agree more with the many views on poor workmanship and bad after-sales service that have been published in Practical Wireless. It's about time that high standards of the above were adopted as company policy to ensure value for money.-Ian Vine, (Middlesex).

test bench amp E.BUCKLAND



NYONE who has more than a passing interest in the construction of radio and electronic projects will have acquired a fair number of items of test gear. A multimeter is essential but an r.f. and a.f. signal generator will also be found very useful. These items, plus others which will be required for one's own particular field of interest, are used mainly to test and trouble-shoot finished or partially finished equipment.

One of the most useful items in the author's opinion, however, is none of the above but a simple straightforward audio amplifier; this is rarely considered as an item of test gear. This is surprising when one considers how useful it can be.

A high proportion of the projects published in this magazine are fitted with an amplifier as the final stage; a quick check shows that 30 of the last 36 projects featured on the cover of *P.W.* either used an audio amplifier or were designed to feed one. This gives a rough indication of how often a test bench amplifier could be used for checking that either the early stages of such a project are operating properly or even that an amplifier is working correctly. The circuit shown here can equally well be used as a signal tracer at audio frequencies.

To be of maximum use such an amplifier has to meet certain requirements. Portability was considered important and so the circuit is battery operated. Small size was also high on the list of priorities as this will mean less shelf space and will make it easy to carry around in a brief case for instance. It had to be of reasonable quality, at least good enough to be able to distinguish between correct and incorrect operation of the equipment to which it is coupled. High input impedance and high sensitivity are also desirable features for such a circuit. The output level is of less importance as long as it is to be used mainly as a monitor. The output is in the order of 200mW but small speakers are not very efficient and this output is less than it sounds. Even so, this compares favourably with the output from normal small transistor radios using the same type of battery and has been found to be more than adequate for the intended purpose. In fact many of the design features were controlled by the use of a PP3 battery. This is used to meet the requirements of portability and small physical size. The permissible current drain from this battery limits the output to the level mentioned above.

All the components used are widely available and the cost of this project is not high—certainly not over £2 in total.

THE CIRCUIT

The circuit of the Test Bench Amplifier is shown in Fig. 1. The input is applied directly across the volume control VR1 which is rather higher in value than one would normally encounter in a circuit of this type. This high value has the advantage here of presenting a high impedance input at low volume control settings—this approaches $500k\Omega$, though this is somewhat reduced for high volume settings where

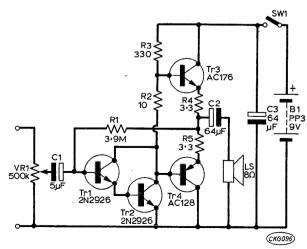


Fig. 1: The complete circuit of the Test Bench Amplifier.

the input impedance of the first transistor is in parallel with the volume control.

The slider connects to the base of the first transistor via the d.c. blocking capacitor C1. Trl and Tr2 are connected as a Darlington Pair, a configuration which gives very high gains and also a higher input impedance than a conventional commonemitter amplifier would present.

This stage drives the output pair Tr3 and Tr4. These are conventionally connected as a complementary output pair; one is an NPN and the other a PNP with the bases connected together by R2, a 10Ω resistor. This is necessary to provide a small bias to avoid cross-over distortion.

When Tr2 is conducting heavily (when it is driven on by a positive swing in the signal), Tr4 is driven into conduction. When Tr2 approaches cut-off (due to a negative going swing) the voltage at the base of Tr3 rises (R3 providing the bias) and it conducts. This explanation is far from complete but descriptions of the operation of this type of output stage are described more completely from time to time in other articles.

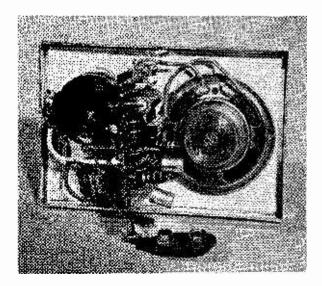
Note that Tr3 and Tr4 must be a matched pair of transistors but are not otherwise critical. Any similar complementary pair of germanium transistors can be substituted without any circuit modifications.

The bias for the Darlington Pair is provided by R1 which also introduces both a.c. and d.c. feedback. If difficulty is experienced in making the circuit work, a slight change in this value may be helpful. However four prototypes have been built of this circuit and this value has proved satisfactory in each case.

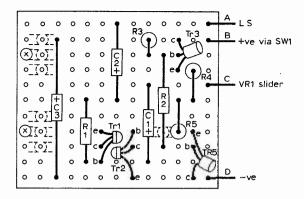
The low value of C2 certainly reduces the bass response but this makes very little difference in practice—miniature loudspeakers are not normally renowned for their good bass response; there seems little point in providing a signal which cannot be handled by the speaker. This smaller value than one would normally find has the added advantage that it is physically small.

CONSTRUCTION

The majority of the components are mounted on



An internal view of the prototype.



- Mounting holes
 Indicates break in copper strip
- CKG097)

Fig. 2: The component layout on Veroboard.

* components list

3.9M Ω R4 3:3 O R2 10 Ω R5 3·3 Ω R3 330 Ω All ¼W, 10% types VR1 500k Ω log, pot, with switch Capacitors C1 5μF 10V C2 64μF 10V C3 64µF 10V Transistors Tr1 2N2926 AC1761 Tr2 2N2926 Tr4 AC1981 † Matched Pair. Any "colour" of 2N2926 transistor may be used for Tr1 and Tr2. Miscellaneous L.S. Miniature 8 Ω loudspeaker PP3 battery, 9V Battery clip; Veroboard 0-15in. matrix, 13 x 11 holes; metal or plastic case-see text; Jack plug and socket; Crocodile clips.

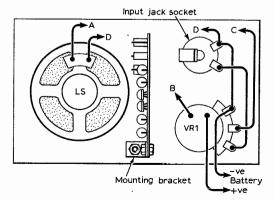


Fig. 3: The wiring details, The wires marked A to D connect to the appropriate points on the circuit board.

a small piece of Veroboard, 0.15in matrix, 13×11 holes. One end is left free of components to leave room for a small mounting bracket. The components layout is shown in Fig. 2. The copper conducting

strip has to be cut away around the mounting holes for obvious reasons but other than this only one additional break is needed.

Due to the low output and the protection afforded the output pair by R4 and R5, heatsinks are not really necessary on Tr3 and Tr4 when operating from the specified battery.

The circuit can be fitted into almost any small case. The author originally built his into a plastic case but this was dropped several times and eventually broke. For this reason it was replaced by a small metal case which proved much more satisfactory. The one used in the prototype was 4×2^{1} ₂ x 2in. fitted with a lid; this is available by mail order from Henry's Radio.

The lid carries the components and a wiring diagram for the final connections is shown in Fig. 3. Of course, any convenient case can be used. The loudspeaker is glued to the face; this method of fixing has been found to be very simple and quite strong enough.

The input to the amplifier is through a jack socket and two wires from the associated jack plug can be fitted with croc clips for rapid connection to any circuit. A band of tape was wound around one of these wires to identify the "earthy" connection.

Proper distortion and frequency response figures have not been taken from this circuit but the amplifier introduces no noticeable distortion and the frequency response has proved to be quite adequate for the designated purpose.



BOOKS WANTED

BOOKS WANTED
...Mullard Circuits for Audio Amplifiers. Second Edn. 1962.—George W. Saunders.
...Any volumes 3 to 10 "Wireless World", RSGB T & R Bulletin, July 1933 (with covers), RSGB 1928 Annual Log Book. "History of Radio Telegraphy & Telephony". Blake, 1927.—Alian Herridge, G31DG, 96 George Street, Basingstoke, Hants.
...Oscilloscope equipment circuits by Easterling buy or borrow.—F. Cosgrove, Rowan Chalet, Lower Road, Postcombe, Oxford, OX9 7DU.
..."Simple Radio Circuits" by A. T. Collins.—J. Wheelton, 41 Chorley Road, Burntwood, Walsall, Staffs.
...Young Schoolboy requires a second-hand copy of the Radio Communication Handbook about £1? Phone Harpenden 5910.—Huw Hallybone, 78 Sauncey Avenue, Harpenden, Herts.

INFORMATION AVAILABLE

IRANS: 9.222-18, REC: 9.209-R.—R. Peel, 57 Tangmere Drive, Castle Vale, Birmingham, 35.

EXCHANGE
...P.E. April, 70 to Nov. 70 & March 71 to Dec. 71. Also P.T. Jan. 71 to Oct. 71.
27 issues. All good condition. Would exchange all these for Vol. 43 & 44 P.W.—
G. J. David, 32 Trebeferad, Llantwit Major, Glam.
... have the Aug. 1966 edition of P.W. twice so am willing to exchange one copy for any interesting bit of electronic junk—L. Cook, 7 Plum Tree Close, Eccleston Park, Prescot, Lancs., L35 7JT.
... "Swap, buy or sell P.W. back issues. Large quantity for disposal. S.A.E. for lists,"—R. Forsbert, 123 Harestone Hill, Caterham, Surrey, CR3 6DL.
...June 1968, July and Aug. 1968 issues of P.E.—S. Tucker, 8 Hawkswell Gardens, Oxford, OX2 7EX.
...May 1971 issue of P.W.—C. H. Cheah, 17 Hargrave Road, Archway, London, N19 5SH.
...Aug. 1968 P.W. or photocopy of Portable Keyless Organ.—G. E. Dutton, Buck-tith-Vine-Inn, Burscough Street, Ormskirk, L39 2EG, Lancs.
...P.W. Volume 45 (May 59-April 70), P.W. Volume 47 No. 7 (Nov. 71), P.E. Volume 5 (Jan. 69-Dec. 69 inc.).—M. Ridger, 25 Broomfield Crescent, Leeds, L55 3DD.
...P.W. for Jan. 1971 with Part 1 of the Stereo Tape Recorder.—R. W. Andrews, 7 Bracken Grove, Wellington, Salop.
...The issue of FW. Containhing bloom is Rullon Road, Peniculk, Midlothian.
...The bloom of the P.W. 35W Gultat Amplifier.—P. Winkley, 38 Leopold Avenue, Handsworth Wood, Birmingham, 20 .
...July 1971 issue and any 1989 issues of P.W.—N. A. Dent, Middle House, Lockner Holt, Chilworth, Guildford, Surrey.

CHARLES MOLLOY

OMMY CROSBIE who lives in Broseley, Cheshire writes "I do not think a communications receiver is necessary for a beginner. I only use a domestic receiver (removed from an old radiogram) though some modifications must be made." He then describes how he traced the a.g.c. line and connected it to chassis via a switch. When the switch is operated the a.v.c. is off and the receiver is ready for DXing. Stations heard by Tommy on his modified receiver and MW loop include RNE Tenerife, Canary Islands on 620kHz; Godhavn, Greenland 650kHz; PJB TransWorld Radio, Bonaire 800kHz; EFJ57 Tenerife 890kHz; WINS New York 1010kHz; WBMJ San Juan, Puerto Rico 1190kHz; Radio Afghanistan, Kabul 1280kHz; PJD St Maartin, Netherlands Antilles 1295kHz (this station often broadcasts in English).

DXers will find it an advantage to switch-off the a.v.c. when operating on the medium waves. Two or more stations can often be heard simultaneously on a channel and the slight difference in frequency between them gives rise to a beat. If the a.v.c. is switched on, then the receiver gain will follow this beat, giving rise to an unpleasant flutter which makes reception difficult. The writer invariably has the a.v.c. switched on, then the receiver gain will follow this while the receiver gain is adjusted with the r.f. gain control.

Paul Swain, while on a visit to Tudweilog on the coast of Wales, used his Sanyo 7 transistor portable between 0003hrs and 0330hrs GMT on the 12th February to log CBN St John's, Newfoundland 640kHz; WOR New York 710kHz; CJON St John's 930kHz; WINS New York 1010kHz; WNEW New York 1130kHz; WWVA Wheeling, West Virginia 1170kHz; WHAM Rochester, New York 1180kHz; WOWO Fort Wayne, Indiana 1190kHz; WCAU Philadelphia, 1210kHz; WTOP Washington, D.C. 1500kHz; WMEX Boston 1510kHz; WKBW Buffalo, N.Y. 1520kHz.

Richard Coyle of Glasgow sends a log heard on a Lafayette KT340 and an antenna of 200ft of wire wound round the loft in a triangular coil. R. Caracas, Venezuala 750kHz; WHDH Boston 850kHz; WCBS New York 880kHz; CBM Montreal 940kHz; R. Sutatenza, Colombia 960kHz; WBZ Boston 1030kHz; WHN New York 1050kHz; Radio Globo, Rio 1180kHz; R. Tupi, Rio 1280kHz.

A new high power outlet at Bissau, Portuguese Guinea has appeared on 1070kHz. According to a OSL received by the writer it is on the air daily until 0100hrs GMT and has been heard as a strong signal with deep fading, after 2300hrs. The address for reception reports (which may be in English) is Emissora da Guine Portuguesa, Caixa Postal 191, Bissau, Portuguese Guinea.

Send logs and information about the Medium Waves to the writer at 132 Segars Lane, Southport, PR8 3JG.



S. GINSBERG

HINGS are hotting up in the television field. Hitachi have launched a 20in. colour tube with 110° deflection angle and with a 29mm neck. The Japanese market version of the tube will have a black matrix but because of Zenith patents the one for Europe will not. Hitachi have plans to follow this tube up with smaller versions in 18in. and 16in. One interesting thing about the Hitachi tubes is a change in the approach to avoid a form of distortion.

A problem with many 110° tubes is that the static convergence at the edges of the tube is upset because of the large convergence angle near the edges of the tube. The electrons passing through the aperture in the shadowmask do not land accurately on the corners of an equilateral triangle and thus colour purity is degraded. While some people might claim that corrective action during tube manufacture tends to eliminate this defect, Hitachi claims that these remedies are inadequate.

Hitachi are fabricating tubes to match this electron distortion rather than trying to correct the placing of the electron beam itself. The resultant tube permits components such as those used with 90° tubes to be used rather than require special deflection coils and special components.

Back to computers. Memories are very much in the news but perhaps the most interesting one is not a semiconductor memory nor a magnetic core. It is a ferroelectric one which, if researchers are right, will pack some 1013 bits of information. The idea came originally from holograms. These are three-dimensional images. It was reasoned that perhaps information could be stored in three dimensions, one layer behind the other. Now comes the idea that this could be done with ferroelectric materials and the suggestion is for one called barium titanate which has been doped with impurities. The doping causes the individual crystals to become photosensitive besides being transparent. Thus they could be used to record the interference waves from a holographic image. They could easily be "read out" by using a beam of light and they can also be erased and new information "written" in. So it looks like another twist in the memories stakes with a serious competitor for the memory elements; semiconductor, hologram, plated wire, magnetic core

RCA are recorded as having done some work on ferroelectric crystals but this method used a great deal of heat (some 300°C required for erasure) and there were few problems. The newer method just announced, using barium titanate, uses only an electric filed for erasure and thus looks promising. Experiments, incidentally, are taking place in France.



A components catalogue is so vital to any keen constructor that it simply does not pay to make do with less than the best. True, the best may cost a little more.. but it's the cheapest in the end. So invest in a Home Radio Components Catalogue, listing over 8,000 items, more than 1,500 of them illustrated. If you call at our shop the catalogue is yours for just 50 pence. If you order by post—70 pence, including postage and packing. You also get 10 Vouchers, each worth 5 pence when used as instructed—so you can get the cost of the catalogue back in any case!

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practically wireless HENRY commentary by

No. 90

Nothing

if not

Informative

HATTERED, not a little disabused, Henry sits hunched in the corner of his workshop trying to work out where he went wrong.

No, it is not another classic case of diagnosing power supply, regulator and control circuit malfunction when the heat-fuse in the mains transformer had parted. Not a mere inability to restring a dial drive cord that goes twice round the town hall and back between tuning gang and pointer pulley. Not even, not especially, another case of reading nano for micro or milli for Meq.

Worse than that, Henry has been accused by a close colleague—nay, a collaborator—of being too literary, not factual enough. 'People pick up *PW*,' he says, 'to be informed, not to be led through the pages of the dictionary.'

Oh dear—I plead guilt. Mea culpa, and all that. Verbal diarrhoea has afflicted your scribe.

There—you see? Put into plainsong, the foregoing should read: Henry talks too much, and to little effect.

'Tis true, 'tis true. Which is why Henry, in this lazy month between Spring and Summer, wants to acquaint you with a few of the



Large frightening sparks.

facts he has stumbled upon during a recent browse through the trade and enthusiast magazines.

In the April 1972 issue of Studio Sound there are two equipment reviews. First is of a Crown DC 300 amplifier, imported from Indiana, U.S.A. A cool 600 watts into 4 ohms, is all. Or, to be a little more realistic, 150W RMS per channel into 8 ohms. The reviewer, P. A. Lomas, had little to say about it except that it exceeded specification' for every test made. And, believe me, those specifications are very impressive, as one would expect for a bit of hardware costing £360.

What interests Henry more is the methods he used to determine how good this amplifier really was. . . . Harmonic distortion of 0.008% at $500 \mathrm{Hz}$, Crosstalk at $10 \mathrm{kHz}$, $95 \mathrm{dB}$ below full output, and noise at $-113 \mathrm{dB}$, forsooth!

He does report—and Henry applauds the touch of humanity—'Short circuit tests merely produced large frightening sparks, pitted screwdrivers and a shaking hand.'

But the accompanying review was rich. It dealt with the Edison Phonograph, and could have been better if a perfectly straight-faced approach had been maintained. Instead, we read such specifications as:—Wow and Flutter: dependent on alcohol level in blood of operator. (The Phonograph is a cylindrical-scan tinfoil clad machine operated by a hand crank, or hadn't you guessed?)

In fact, the review goes into nice detail about the construction, with diagrams, and nowhere does T. T. Wittering mention April 1. If David Kirk, the Editor of Studio Sound, intends to continue with the April Fool insert, as Radio-Electronics used to do (still does?), T. T. W. will have to get together with Henry Scruggs or George Izzard O'Veering and write about the Super-Crown.

On second thoughts, he need only quote from some specifications as boldly published by the



The 'ultimate' Hi-Fi system.

makers of system audio equipment. And if he wants more power, what about the Marantz at 250 watts, or the Phase Linear, 750 watts?

Power alone isn't everything, as any owner of an earache generator can tell you. But it does seem to be the tendency for makers of powerful amplifiers to make, also, equipment that performs to the highest standards.

Bert Whyte's piece in the February Audio talks about Joe Audiophile, in his seventh heaven because his Aunt Nelly remembered him in her will. He purchases the 'ultimate' hi-fi system, super megawatt amplifiers, preamps with a plethora of controls, which can be corrective or creative, digital readout tuner—naturally, Joe's system is quadrophonic—and ultra-wide range speakers with low frequency response down in the subbasement.

Developing his theme, Bert describes that Joe's aim is to listen to his 15 ips copies of classical masters. 'The first faint susurations (sic) of Ravel's 'Daphnis and Chloe' are heard from the speakers . . . molto pianissimo . . . and Joe is in a transport of delight.' no tape hiss —but then—WHUMP! RUMBLE'

Joe has been the victim of monitoring techniques. The discs made from those master tapes have been 'rolled off rapidly' below 60Hz.

So many of the components you need for PW designs are in the new 1972 Electrovalue catalogue. Bigger, better than ever-Post free-10p.

ELECTROVALUE

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	Ger. NPN		26p
	PNP	11	26p
	Sil. UJT	Oscillator, SCR driver	47p
	NPN	Small sig. amp	11p
	NPN	High power	50p
	PNP	Low power	10p
	NPN	Low power	10p
	Ger. PNP	Small sig./driver	23p
	PNP	Low power	20p
	NPN	Low power	16p
	PNP	High power	58p
l	NPN	Med. power	83p
	PNP	Med. power	36p
	Sil. NPN	Small signal	11p
	NPN	Low noise	12p
	NPN	Small signal	10p
	NPN	Low noise	11p
	NPN	RF amp.	14p
	NPN	Med. current	20p
	Ger, diode	RF detector	6p
		General	5p
		Silicon Rectifier 1 amp	10p
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Code	Power	Tolerance	Range in ohms	Values		10 to 99 note belo	
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Codes: C = carbon film high stability low noise

MO = metal oxide Electrosil TR5 uitra low noise WW = wire wound Plessey

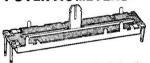
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Values: Values: 10, 12, 15, 13, 22, 27, 33, 38, 47, 56, 68, 82 and their decades. E24: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 52, 72, 91 and their decades. Froces are in penne each for same obnic volue and power rating, NOT mixed values. (Ignore fractions of 1p on total value of resistor order).

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				83p
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			AN)	1.28
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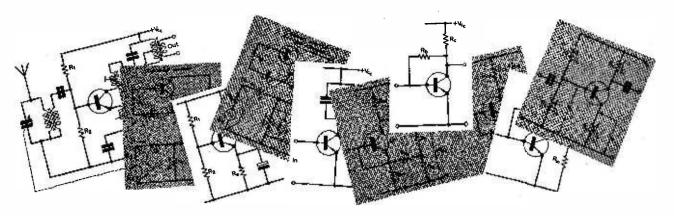
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PART 8

Buffer links

It has become obvious in recent months, from the polite noises made over the editorial hot-line, and correspondence we have received, that our chosen method of dealing with simple transistor circuitry meets with some approval.

Simplicity can be deceptive. This month, for instance, we have chosen as our subject a singletransistor collector-follower circuit. One transistor. a few components, a little bit of board, and a few moments of your time. But if we were to cover, fully, all the parameters affecting the calculated performance of this 'simple' circuit, half the adverts would be squeezed out of PW. More important, such an approach would frighten off the beginnerand he's the chap at whom this series was aimed, remember?

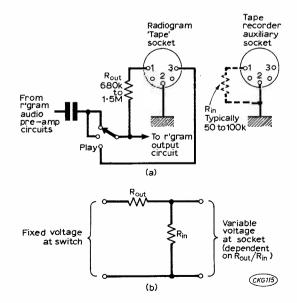


Fig. 39. (a) A single channel of a typical radiogram 'tape' outlet showing the presence of Rout which affects the matching. (b) is the equivalent of (a).

The buffer-link can be used between two pieces of equipment when discrepancies of impedance and voltage are such that gain as well as matching is needed. A typical example is the case of the tape recorder connected to the 'Tape' outlet of a radiogramophone. Quite often, specifications make it appear that the simple connecting lead will do the job adequately. The significant point omitted from those specifications (or, at best, skated over), is that the impedance of the load drastically affects the available signal voltage from the radiogram. A glance at Fig. 39 shows why this should be.

To prevent the load (i.e., the tape recorder) from robbing the main equipment of some signal voltage and to save the extra cost of a properly designed feed stage, such as we have already described, a fairly high resistor is used to connect the signal take-off point to the socket.

If a direct connection is now made to the socket, the relatively low impedance of the tape recorder will not affect the performance of the radiogram. But, and it's a big but, a potential devider is now formed, with the smaller resistance section being the tape recorder, so the available voltage is reduced in proportion to the two resistors.

If there is not enough voltage available, we are not going to be able to make a decent recording. No amount of ingenuity with external resistive networks will produce sufficient modulation. What we need is a little amplifier between the radiogram and the tape recorder. We need, in fact, a buffer to prevent interaction of one upon the other and a link to join their circuits together—a buffer-link.

Simple circuit

Emitter-follower circuits make excellent buffers, but do not allow us to obtain any voltage improvement. A two-stage circuit would do the trick, maybe; like the Darlington Pair of Part 6; or even more elaborate circuitry, like the single-double device of Part 7. Here, we can obtain the gain we require and make a suitable match with the very simple circuit of Fig. 40.

This is the single-transistor collector follower stage. If you have read Part 5, page 913, it should not be necessary for me to explain those terms. Such a stage has a medium impedance input, certainly lower than the emitter follower we previously discussed, but still not too low for our purpose. It has a fairly high output impedance, but again, not too high for our purpose. Component changes can modify the performance to suit our requirements, as we shall show. Voltage gain is quite reasonable, and the signal inverts 180° between input and output. All these are conditions that we want.

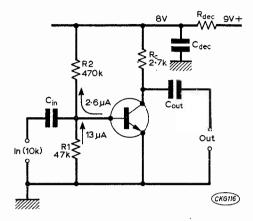


Fig. 40. Basic buffer-link circuit. Essentially a collector-follower the emitter is taken directly to the negative line.

R1 and R2 form the potential divider to give us base bias, just as we have seen in previous months. $R_{\rm c}$ is the collector load resistor, and this time, our a.c. signal is developed across this load. The emitter in our first example is taken directly to chassis.

The input signal to the base is coupled by $C_{\rm in}$, with $C_{\rm out}$ performing its coupling function at the other end of the stage, taking the signal to the input of the tape recorder.

Recap

Recapping a little: previous dealings with the emitter follower and Darlington circuits have taught us that the output signal is a little less than the signal voltage fed in to the circuit. In technical terms, the gain is less than unity. This time, we have acquired a bit of gain, and the price we pay is an input impedance lower than before, and an output impedance higher than before. In addition, we now have a phase inversion, which the emitter follower did not have. A signal at the input of the stage is inverted, i.e., receives a phase change of 180°. If the input signal is positive-going, the output signal will be negative-going. This doesn't much matter to us in our present application, but can be quite important for some applications.

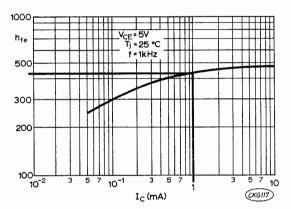
On the subject of impedance, the terms high, low and medium are, of course, only relative. Impedance is the resistance to an a.c. signal, usually at a specified frequency. Where no frequency is specified, as in the case of an amplifier, it is assumed that the impedance is constant over the frequency range of interest: example, 20Hz to 20kHz. A reference frequency of 1kHz is generally assumed, except in the case of microphones and loudspeakers, when 400Hz is more often employed. In general, an a.c. resistance of 1k Ω and below would be called low, up to 100k Ω would be medium and above 100k Ω referred to as 'high impedance'.

Circuit details

We should begin with some modest requirements, bearing in mind that a wide variation of performance can be expected with certain circuit changes. Battery supply—again 9 volts, allowing for 1 volt being dropped across the decoupling resistor, so the stage supply $V_{\rm cc}$ is 8V. The collector current we shall choose is 1mA. And with these simple starters we can choose a transistor.

Our choice is determined by the need more for low noise than high gain, although it is nice to have both. So, to get the best of both worlds, back to old faithful, BC109. Referring to Figs 41 and 42, we can work out the more detailed figures.

The transistor will be operating at a collector current of 1mA, and the voltage between collector and emitter, V_{GE} , will be about 5 volts. (Refer to previous articles in this series for reasons for these quoted figures—an expansion of the argument). From Fig. 41, we can see that the a.c. current gain. h_{fe} , can be determined at somewhere around 450. Similarly, from Fig. 42, we find that the d.c. current gain, h_{FE} , is around 380.



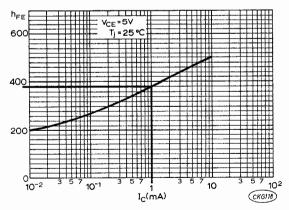


Fig. 41. (top) Graph to determine h_{fe}. the a.c. current gain. Fig. 42 (bottom) is the graph to find the d.c. current gain h_{Fe}.

The base current of the transistor is determined by the formula:

$$I_b = \frac{I_c}{h_{FE}}$$

We have determined $h_{\rm FE}$ and the I_c we chose, for best conditions and easy working, to be 1mA. Putting these figures in the above formula, we arrive at the conclusion $I_b = 2 \cdot 6 \, \mu A$.

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RADIO SOCIETY OF GREAT BRITAIN 35 DOUGHTY STREET, LONDON, WC1N 2AE Referring again to our previous arguments, we say that the current flowing in the base bias chain, R1 and R2, should be about five times the base current of the transistor. In our circuit, we have the emitter taken directly to the negative line, and we know that the base-emitter voltage of a silicon transistor is normally 0.6V. So we can calculate the value of R1, since the voltage across it must be 0.6V and the current through it five times the base current of $2.6\mu A$. If we do our sums correctly, the answer will be:

$$R1 = \frac{0.6V}{5 \times I_b(2.6\mu\text{A})} = \frac{0.6 \times 10^6}{13} = 46,154\Omega.$$

The nearest preferred value in the 5% range will be $47k\Omega$.

The voltage across R2 will be the supply voltage, 8 volts, minus the base voltage of 0.6V = 7.4V. The current will be six times I_b , that is the current in R1 plus the original base current. So:

R2 =
$$\frac{\mathbf{V}_{cc} - \mathbf{V}_{b}}{6 \times \mathbf{I}_{b}} = \frac{8 - 0.6}{15.6 \times 10^{-6}} = \frac{74 \times 10^{6}}{156} = 474,359\Omega.$$

The nearest preferred value is $470k\Omega$.

Input resistance

We have talked before about input resistance and our own experiences in trying to match equipment, about which manufacturers have given inadequate information, show that this is a difficult field. 'Fings ain't always wot they seem ter be!'

The input resistance of the transistor (looking into its base and ignoring R1 and R2) is $h_{\rm ie}$. Fig. 43 shows that Mullard graph for the BC109, idealised for the conditions under which we are using the transistor—that is, with unnecessary information omitted. We have done this here, at Michael's insistence, because, to the layman, there is nothing more confusing than a graph filled with curves and references he is not called upon to use.

Here, we have indicated that the h_{ie} is a little over $10k\Omega.$ If we had no graph, we should have to calculate:

$$h_{ie} \!=\! r_e \! imes \! h_{fe}$$
 where $r_e \! = \! \frac{25\Omega}{I_c \ (mA)}$

so $h_{ie} = \frac{25}{1} \times 450$ or $11.25k\Omega$, which is reasonably near

the plotted figure in this case.

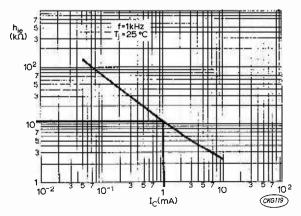


Fig. 43. This graph will give the input resistance, h_{ie} , of the transistor.

The input resistance of the stage, and not just the h_{ie} , is the latter shunted by the parallel combination of R1 and R2. This is as far as a.c. is concerned. (In parallel, because to a.c. signals, by reason of the low impedance of C_{dec} , the top of R2 is effectively connected to the bottom of R1.)

So we calculate for hie, R1 and R2 in parallel, arriving at:

$$\frac{1}{11k\Omega} + \frac{1}{47k\Omega} + \frac{1}{470k\Omega} = 8.7k\Omega \text{ approx.}$$

If there are to be 5 volts between collector and emitter, then we must have a voltage drop across $R_{\rm c}$ of 8-5 ($V_{\rm CC}-V_{\rm CE}$). This works out to 3 volts and if the collector current is chosen to be $1 {\rm mA},$ the resistance of the load $R_{\rm c}$:

$$\frac{3 \text{ volts}}{1 \text{mA}} = \frac{3,000}{1}$$

The nearest preferred value to a $3k\Omega$ resistor we shall get in the 5% range is $2\cdot7k\Omega$. That's near enough.

Gain

Stage voltage gain (A_r) is equal to R_c divided by the emitter resistance. In this case we are concerned with the **effective internal** emitter resistance, r_c , which you will remember, caused some confusion earlier, and is calculated from the formula:

$$r_{e} = \frac{25}{I_{c} \text{ (mA)}} = 25\Omega.$$
 The stage gain, A_v, becomes:
$$\frac{R_{c}}{-} = \frac{2,700}{-} = 108$$

We made some comment on the principle of selecting transistors so this may be the place to underline the importance of parameter variations on the validity of some associated calculations. In this case, if I_{\circ} changes, so does r_{\circ} . We chose our collector current and took a 'typical' $H_{\rm FE}$ from the graph. But collector current (actual) depends on base current and actual $h_{\rm FE}$. So if the selected transistor has an $h_{\rm FE}$ that differs, $I_{\rm C}$ will differ, $r_{\rm e}$ will be affected and the stage gain may be quite different from that calculated. Hard world, isn't it?

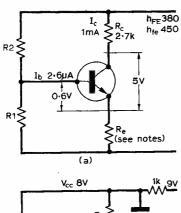
Adding an external R.

In the previous case, that of Fig. 40, we had the emitter firmly strapped to 'earth' so the emitter voltage was the same as the negative supply line. What happens to the stage and its performance if we now insert a resistance in this emitter, in the position $R_{\rm e}$ of Figs 44 (a and b)?

Quite simply, the stage gain is altered and at the same time the input resistance, $R_{\rm IN}$, is increased.

Basing our calculations on some of the factors we have already, we can choose a convenient value of resistor for R_e and work out what differences it will make. First, V_{oc} is 8V; I_e , ImA; R_e , $2 \cdot 7k\Omega$ and our R_e will be, let us say 47Ω . Then $V_e = I_e \times R_e = 1mA \times 47$ ohms = $0 \cdot 047V$.

The emitter current is the collector current plus base current, or $I_e = I_c + I_b$. Since the base current of the transistor (2.6 μ A) is very small in comparison with the collector current of lmA, it is convenient to ignore it, satisfied that there will be only a negligible amount of error. So we can say, effectively,



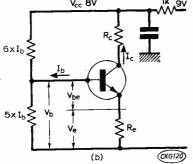


Fig. 44. (a) Theoretical collector follower circuit, showing voltage and current distribution. (b) shows the symbolic representation with R1 and R2 values indicated relative to I_b.

 $\begin{array}{l} V_b\!=\!V_e\!+\!V_{be} \ (\mbox{Note:---the base-emitter voltage of a silicon transistor, when it is forward biased, as in normal operation, is about 0.6V). Thus, <math display="block">V_b\!=\!0.047 + 0.6 =\!0.65V, \mbox{ which is also the voltage across } R1. \end{array}$

R1=
$$\frac{\dot{V}_b}{5 \times I_b} = \frac{0.65}{5 \times 2.6 \times 10^{-6}} = 50 kΩ$$
, approx.

To the nearest preferred value in the 5% range, we can choose $47k\Omega.$

$$R2 = \frac{V_{cc} - V_b}{6 \times I_b} = \frac{8 - 0.65}{15.6 \times 10^{-6}} = 471.154 \text{k}\Omega$$

Again, we choose the nearest preferred value, and settle for $470k\Omega$ in the 5% range.

Having added $R_{\text{e}},$ we now have to take account of its presence in the voltage gain formula, $A_{\text{v}}\colon$

$$\frac{R_{c}}{(R_{e}+r_{e})} = \frac{2,700}{47+25} = 37.5$$

Input impedance

We were content to deal with $h_{i\sigma}$ previously, the input resistance (impedance) of the transistor, looking into the base with R1 and R2 ignored. Now that we have added R_{e} , we must allow for it in calculating the input impedance, and we shall be dealing with R_{in} :

 $R_{\rm in} = (R_{\rm e} + r_{\rm e})(h_{\rm fe} + 1) = (47 + 25)(450 + 1) = 32 \cdot 472 k\Omega$ The shunting effect of R1 and R2 has to be considered again, as far as a.c. is concerned, so the stage input resistance $R_{\rm IN}$ is calculated as before:

$$\frac{1}{R_{\rm IN}} = \frac{1}{R1} + \frac{1}{R2} + \frac{1}{R_{\rm in}} = \frac{1}{47k\Omega} + \frac{1}{470k\Omega} + \frac{1}{32k\Omega}$$
= $18 \cdot 3k\Omega$, approx.

Negative feedback enters into matters at this point. We have not fitted any bypass capacitor across

the emitter resistor. This has allowed series negative feedback to take place, and the gain of the stage is reduced by the negative feedback. As it effectively changes the value of R_0+r_0 , the stage input resistance is also affected. This feature is widely used in transistor circuitry, where negative feedback is employed to set the stage gain, alter the input impedance, improve the frequency response and reduce harmonic distortion.

To demonstrate the differences that are obtained when $R_{\rm e}$ is inserted, and then altered, the accompanying table has been prepared. Calculations which back these figures are based on the foregoing formulae, as worked out for an $R_{\rm e}$ of 47Ω , which we have used. We are assuming a $V_{\rm co}$ of 8V, and $R_{\rm e}$ of $2.7k\Omega$, using a BC109 transistor. We have taken the figures calculated and measured for no $R_{\rm e}$ and for four different resistive values.

$R_e \Omega$	0	47	100	220	470
Output impedance R1 ($k\Omega$) R2 ($k\Omega$) R2 ($k\Omega$) measured R $_{iN}$ ($k\Omega$) calculated	<2·7k 47 470 8·6 8·7	2·7k 47 470 19·5 18·3	2·7k 47 470 26 25	2·7k 56 470 36·3 34·3	2·7k 82 390 52 52
A _v (voltage gain) measured A _v calculated	120 108	35 37·5	19 21·6	10·5 11	5·4 5·74

Table indicating the changes in the circuit characteristics with different values of R_e.

Finally, a paragraph on coupling and decoupling. We have already stated that R_{dec} is dropping 1V and we know that the approximate current through it will be lmA so from this we can calculate its value,

$$R_{dec} = \frac{1V}{1mA} = 1k\Omega$$

We can take it that the decoupling capacitor, C_{dec} is as before, that is $100\mu F$. C_{out} is similar, at $10\mu F$ and C_{in} will vary with the stage input resistance for optimum value. Using the rule-of-thumb method, $1\mu F$ into $100k\Omega$, and applying this to measured values of $R_{\rm IN}$, we can see that this will vary from a mere $0.1\mu F$ (actual value $0.086\mu F$) to greater than $0.5\mu F$.

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9G302 90p 2N3405 45p 40311 35p BCV30 27½p BSX60 82½p NKT401 2G306 42½p 2N3415 22½p 40312 47½p BCV32 50p BSX76 22½p NKT402 2G308 30p 2N3415 32½p 40314 37½p BCV32 50p BSX76 22½p NKT403 2G309 30p 2N3417 37½p 40820 47½p BCV32 32p BSX76 22½p NKT403 2G371 15p 2N3576 41.25 40934 47½p BCV38 40p BSY76 27½p NKT406 2G374 20p 2N3577 97½p 40826 37½p BCV38 40p BSY70 27½p NKT405 2G374 20p 2N3572 97½p 40826 37½p BCV38 40p BSY10 27½p NKT405	TRANSISTORS						
9G302 20p 2N3405 45p 40311 35p BCV30 271p BSX60 82ip NKT401 2G306 42ip 2N3415 22ip 40312 47ip BCV32 50p BSX76 22ip NKT402 2G308 30p 2N3415 32ip 40820 47ip BCV32 50p BSX76 22ip NKT403 2G309 30p 2N3417 37ip 40820 47ip BCV32 30p BSX76 22ip NKT403 2G371 15p 2N3570 41.25 40934 47ip BCV38 40p BSY76 27ip NKT406 2G374 20p 2N3570 41.25 40926 37ip BCV38 40p BSY70 27ip NKT406 2G374 20p 2N3570 27ip 2N26 37ip BCV38 40p BSY10 27ip NKT405	27 1 p						
26303 20p 2N3414 221p 40312 471p BCV31 30p BSK61 621p NKT402 2G308 431p 2N3415 221p 40314 371p BCV32 50p BSK76 221p NKT403 2G308 30p 2N3416 371p 40820 471p BCV33 25p BSK77 271p NKT403 2G371 15p 2N3570 21.25 40932 247p BCV34 30p BSK76 271p NKT405 2G371 15p 2N3570 21.25 40924 471p BCV38 40p BSY10 271p NKT405 2G374 20p 2N3572 971p 4026 371p BCV39 40p BSY10 271p NKT405	87 ap						
2G306 42 jp 2N3415 22 jp 40314 37 jp BCY32 50p BSX76 22 jp NKT403 2G308 30p 2N3417 37 jp 40323 32 jp BCY34 30p BSX76 27 jp NKT405 2G371 15p 2N3576 21 25 40832 47 jp BCY34 40p BSY10 27 jp NKT405 2G371 20p 2N3572 27 jp 40323 32 jp BCY34 40p BSY10 27 jp NKT405 2G374 20p 2N3572 27 jp 40262 37 jp BCY38 40p BSY10 27 jp NKT405 NKT405 20p 2N3572 27 jp AU362 37 jp BCY38 40p BSY10 27 jp NKT405	90p						
2G308 36p 2N3416 37jp 40820 47jp BCV33 25p BBX77 27jp NKT404 2G309 30p 2N3417 37jp 40323 32jp BCV34 30p BSX76 27jp NKT405 2G371 15p 2N3570 21.25 40934 47jp BCV38 40p BSY10 27jp NKT405 2G374 20p 2N3572 97jp 4026 37jp BCV39 60p BSY11 27jp NKT451	75p						
2G371 15p 2N3576 £1.25 40324 47kp BCY38 40p BSY10 27kp NKT406 2G374 20p 2N3572 97kp 40326 37kp BCY39 60p BSY11 27kp NKT451	62 <u>t</u> p						
2G374 20p 2N3572 97\p 40326 37\p BCY39 60p BSY11 27\p NKT451	75p						
2G374 20p 2N3572 97½p 40326 37½p BCY39 60p BSY11 27½p NKT451 2G381 22½p 2N3605 27½p 40329 30p BCY40 50p BSY24 15p NKT452	62}p						
2G381 22kn 2N3605 27kn 40329 30n BCY40 50n BSY24 15n NKT452	62 <u>1</u> p						
	62}p						
2N404 22½p 2N3606 27½p 40344 27½p BCY42 15p BSY25 15p NKT453							
2N696 20p 2N3607 22ip 40347 57ip BCY43 15p BSY26 17ip NKT603F							
2N697 17p 2N3702 11p 40348 52 p BCY54 32 p BSY27 17 p NKT613F							
2N698 25p 2N3703 10p 40360 42p BCY58 22p BSY28 17p NKF674F	80p						
2N706 12+p 2N3704 11p 40361 47p BCY59 22p BSY29 17+p NKT677F							
2N705A 12 p 2N3705 10p 40362 57 p BCY60 97 p BSY32 25p NKT713	25 p						
2N708 15p 2N3706 09p 40370 32p BCY70 20p BSY36 25p NKT781	30p						
2N709 624p 2N3707 11p 40406 574p BCY71 25p BSY37 25p NKT10419							
2N718 25p 2N3708 07p 40407 40p BCY72 17p BSY38 22p NKT10439							
2N726 30p 2N3709 09p 40408 521p BCZ10 271p BSY39 221p	37∮p						
2N727 30p 2N3710 09p 40410 62 p BCZ11 42 p BSY40 32 p NKT10519							
2N914 171p 2N3711 12p 40467A 571p BD116 £1-121 BSY51 321p	32 <u>1 p</u>						
2N916 171p 2N3715 £1 25 40468A 35p BD121 65p BSY52 321p NKT20329							
	47∳p						
2N929 221p 2N3791 £2:06 AC107 30p BD124 60p BSY54 40p NKT2033	<u> </u>						
2N930 272p 2N3819 35p AC126 20p BD131 75p BSY56 90p	37 ≟ p						
2N1090 221p 2N3823 971p AC127 25p BD132 85p BSY78 471p NKT8011							
2N1091 22 p 2N3854 27 p AC128 20p BDV10 £1.37 BSY79 45p	771p						
2N1131 25p 2N3854A 27 p AC154 22 p BDY11 21-62 BSY82 52 p NKT80115							
2N1132 25p 2N3855 27p AC176 25p BDY17 £1.50 BSY90 57p	97‡p						

2N3894A 271p AC104 2N3855 272p AC104 2N3855 30p AC188 2N3856A 35p AC188 2N3856A 35p ACY18 2N3858 25p ACY18 2N3858 272p ACY20 2N3859 272p ACY20 2N3866 21.50 ACY22 2N3866 21.50 ACY24 2N3877 A 40p ACY40 2N3877 A 40p ACY41 2N3900 A 40p AD140 2N3901 372p AD140 2N3901 372p AD140 2N3903 35p AD150 2N3906 372p AD162 NKT80113 £1-12 NKT80211 92½p BDY18 £1.75 BDY19 £1.971 BDY20 £1.121 121p 421p 271p 751p 27p 55p 40p 2N1303 178p 2N1304 228p 2N1305 228p 2N1306 25p 2N1306 25p 2N1307 25p 2N1309 30p 2N1507 174p 2N1631 35p 2N1631 35p 2N1638 273p 2N1638 273p 2N1638 273p 2N1638 273p 2N1638 273p 2N1638 35p BDV38 971p £1.25 C111 BDY38 BDY60 BDY61 BDY62 BF115 BF117 BF163 NKT80212 C424 NKT80213 921p C495 C426 C428 C744 D16P1 371p 30p 371p NKT80214 921p BF167 BF173

£1.00 25p 47½p 37½p 18p 19p 30p 30p 35p 32½p NKT80215 921p D16P1 D16P2 D16P3 D16P4 GET102 GET113 GET114 40p 871p 40p 30p 20p 20p NKT80216 2N3901 2N3903 2N3905 2N3905 2N3905 2N4058 2N4058 2N4060 2N4061 2N4062 2N4244 2N4285 2N4285 2N4288 2N4288 OC20 OC22 OC23 OC24 OC25 OC26 OC28 OC29 OC36 OC36 20p 20p GET119 GET120 2N2147 AF114 BF185 521p 121p 2N2147 2N2148 2N2160 2N2193 2N2193A 2N2194A 2N2217 BF194 BF195 BF196 BF197 BF198 BF200 BF224 BF225 AF115 AF116 AF117 AF118 AF119 AF124 AF125 AF126 AF127 AF138 AF178 AF178 AF180 AF181 GET873 GET873 GET880 GET887 GET889 GET890 GET896 GET896 30p 20p 221p 221p 221p 2N2218 OC42 OC44 9N9919 2N2220

25pppp 220ppp 22 GETS96 22\p
GETS97 22\p
GETS98 22\p
MJ400 21-07\p
MJ420 21-12\p
MJ421 21-12\p
MJ430 21-02\p
MJ480 97\p
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MI 410 41071
MI 41071 221p 321p 321p 423p 45p 25p 20p 50p 50p 621p 2N5305 2N5306 2N5307 2N5308 2N5309 2N5309 2N5354 2N5355 2N5356 2N5365 2N5366 2N5366 2N5367 2N5457 15p 15p 10p 20p 20p 20p 21p 10p 12p 12p 11p 12p 11p 12p 11p 11p 71844 TIS44 TIS44 TIS44 TIS45 TIS46 TIS46 BC118 BC118 BC121 BC122 BC125 BC126 BC140 BC147 BC148 BC149 BC152 $2N2925 \\ 2N2926$ 23p 20p 23p N2926 Green 14p Yellow 12½p Orange 12½p N3011 30p N3014 32½p N3054 46p N3055 62p N3133 30p N3133 30p TIS47 TIS48 TIS49 TIS50 TIS51 TIS62 TIS60 TIS62 28005 23p 17 p 57 p 30p 42 p 57 p 67 p 27 p 28020 2N3011 2N3014 28102 98103 28103 28104 28501 28502 28503 3N83 3N128 3N140 BC158 BC159 BC160 2N3053 2N3053 2N3054 2N3055 2N3133 BC167 BC168B 27ip NKT241 26p NKT242 25p NKT243 21-85 NKT244 21-80 NKT245 21-45 NKT261 37ip NKT262 17ip NKT264 17ip NKT271 37ip NKT272 45p NKT274 11p 10p 11p 12p 12½p 15p 15p 22½p 10p 09p 30p 25p 25p 25p 20p 30p 17}p 2N3134 2N3135 2N3136 BC168C BC169B 3N140 3N141 3N142 3N143 3N152 R.C.A. 40050 TIP30A 60h
TIP31A 621h
TIP32A 75p
TIP33A £1.021h
TIP34A £2.05
TIP35A £2.90
TIP36A £3.68 2N3390 2N3391 BC169C BC170 BC171 BC172 BC175 2N3391 2N3391A 2N3392 2 \3393 2N3394 2N3402

BC182 BC183

BC116A

22½p 40309 $32\frac{1}{2}p$ BC184 11p BSX27 471p NKT275 Post & Packing 13p per order. Europe 25p. Commonwealth (Air) 65p (MIN.)
Matching charge (audio transistors only) 15p extra per pair. Prices subject to alteration without prior notice.

TTL.	LOGIC	I.C. N	EW	PRICES

1								
l		12-24		1-11				2-24
1	£p	£р		£р	£p		£p	£p
SN7400	0.20	0.18	SN7433	0.80	0.75	SN7472	0.32	0.80
SN7401	0.20	0.18	SN7437	0.64	0.06	SN7473	0.48	0.41
SN7402	0.20	0.18	SN7438	0.64	0.80	SN7474	0.43	0.41
SN7403	0.20	0.18	SN7440	0.23	0.21	SN7475	0.45	0.44
SN7405	0.20	0.18	SN7441 AN	0.87	0.83	SN7476	0.45	0.44
SN7406	0.80	0.75	SN7442	0.85	0.81	SN7480	0.70	0.65
SN7407	0.80	0.75	SN7443	2.88	2.70	SN7481	1.40	1.38
SN7408	0.20	0.18	SN7444	2.86	2.70	SN7482	0.87	0.82
8N7409	0.20	0.18	SN7445	2.50	2.40	SN7483	0.87	0.82
SN7410	0.20	0.18	SN7446	1.90	0.95	SN7484	2.00	1.85
SN7411	0.23	0.21	SN7447	1.00	0.95	SN7485	3.62	3.40
SN7412	0.48	0.46	SN7448	1.00	0.95	SN7486	0.33	0.80
SN7413	0.40	0.38	SN7449	1.00	0.95	SN7490	0.87	0.84
SN7420	0.20	0.18	SN7450	0.20	0.18	SN7491A		1.10
SN7423	0.51	0.47	SN7451	0.20	0.18	SN7492	0.87	0.84
SN7427	0.48	0.45	SN7453	0.20	0.18	SN7493		
SN7428			SN7454				0.87	0.84
	0.80	0.75		0.20	0.18	SN7494	0.87	0.84
SN7430	0.23	0.15	SN7460	0.20	0.18	SN7495	0.87	0.84
SN7432	0.48	0.42	SN7470	0.40	0.38	SN7496	0.87	0.84
ł .								

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PIV 1A 3A 6A 10A	50 8p 15p	100 9p 17p 	200 10p 20p 25p 57½p	400 11p 22½p 30p 65p	600 12p 25p 32ip 77ip	800 15p 27p 85p 861p	1000 20p 30p — 971p	1200 35p £1.25
15 A		57}p	621p	77 l p	90p	97±p	£1.20	£1.57
35 A		80p	90p	£1 00	£1.25	£1 50	£2·50	
1 nonnand 1	9 2000 00	o planti	a anagra	mlation				

DIODES & RECTIFIERS

IN34A	10p	AA119	.7p	BAX16	12½p	FST3/4	22½p
IN914	7p	AA129	15p	BAY18	17½p	OA5	17p
IN916	7p	AAZ13	12p	BAY31	7p	OA10	200
1N4007	20p	AAZ15	12p	BAY38	25p	OA9	10p
IS44	7p	AAZ17	10p	BY100	15p	OA47	8p
18113	15p	BA100	15p	BY103	22p	O 470	72
IS120	12p	BA102	25p	BY122	47}p	OA73	10p
IS121	14p	BA110	25p	BY124	15p	OA79	7p
IS130	8p	BA114	15p	BY126	15p	OA81	8p
IS131	10p	BA115	7p	BY127	17p	OA82	10p
IS132	12p	BA141	17p	BY164	57p	OA90	7p
18920	7p	BA142	17p	BYX10	22p	OA91	7p
IS922	8p	BA144	12p	BYZ10	35p	OA95	7p
18923	12p	BA145	17p	BYZ11	82p	OA200	7p
IS940	5p	BA154	12p	BYZ12	30p	OA202	10p
		BAX13	5p	BYZ13	25n	TIV307	50B

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2½ × 5in	25p	25p
3‡ × 3†in	25p	25p
3½ × 5in	30p	29p
5 × 17in (Plain)		
Vero Pins (Bag o	f 36) 20 p	
Vero Cutter 45p		
Pin insertion Te		and .15
matrix) at 55p.		

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RESISTORS

	1 matt	5%, 1p.	₽W. 1W & 2W
l			E12 Series
		5%, 1p	E12 Series
i.	1 watt	5%, 1½p	
ľ	1 watt	2%, M/O	4p
	1 watt	10%, 21p	¹W & ∳W
		" O O (D-	7704 0

BRIDGE RECTIFIERS

A, PIV		Α.	PIV	
1 100	37p	4	50	60p
1.4 140	57p	4	100	70p
2 50	32p	6	400 50	80p 62p
2 200	41p	ě	200	800
2 - 400	46p	6	400	£1·10

MULLARD C280 M/FOIL CAPACITORS

0.01, €	0.022, 0	033,	0.047	Sp each
0.068,				4p each
	·22, 0 ·33	3		5p each
0-47				91
0.68				119
$1\mu F$				145
$1.5 \mu F$	• •	• •		21 p
2 211F				25r

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2.5 watt 5% (up to 270 ohms only), 7p 5 watt 5% (up to 8.2k Ω only), 9p 10 watt 5% (up to 26k Ω only), 10p

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The new Mark 2 CODAR CR70A general coverage Communications Far East, Russia, USA, etc. etc., using miniature plug in coils. Hi-Gain FET Regen. Det./Pre-Amp A.F./A.F. Module giving full loud speaker output to any external 2/3 ohm speaker. Two separate Tuning Dials, Calibrated Mnz and Degrees. Electrical Bandspread on all bands. Regen. det. receives CW/SSB signals also. Handsome styling with Black Perspex Panel, Steel Cabinet in Charcoal finish with Chrome Trim.

This is a quality CODAR-KIT with full 12 months guarantee. No technical knowledge required, printed circuit construction, instruction Manual with pictorial drawings backed by friendly Help-U After Sales Service if you have any queries.

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Malcolm Connah

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NEWS FOR DX LISTENERS

THE first reporter this month is **Hugh Cocks** of Mayfield in Sussex who has a Unica UNR-30 receiver and a 70 foot long-wire. Hugh sent in a special log of South American stations as follows:

9515 R. Roquete, Pinto, Brazil at 2200.

9530 R. Calendario, Venezuela at 2230.

9595 R. Cultura de Bahia, Brazil at 2130.

9620 R. Novo de Julho, Brazil at 2150.

9635 R. Aparecida, Brazil at 2225.

9665 R. Nac. Brasilia, Brazil at 2230.

9705 R. Maua, Brazil at 2100.

11720 R. Nac. Brasilia, Brazil at 2100.

11785 R. Guaiba, Brazil at 2058.

11795 R. Nac. Rio, Brazil at 2100.

11805 R. Globo, Brazil at 2100.

11865 R. Cl. de Pernambuco, Brazil at 2100.

11875 R. Soc. de Bahia, Brazil at 2100.

11915 R. TV. Gaucha, Brazil at 2115.

11925 R. Bandeirantes, Brazil at 2200.

11950 R. Min. Educaco, Brazil at 2130.

15105 R. Rural Brasileira, Brazil at 2200.

15145 R. Jornal do Comercio, Brazil at 2150.

15155 R. Dif. de Sao Paulo, Brazil at 2200.

15190 R. Inconfidencia, Brazil at 2230.

15415 R. Cl. Ribeirao, Brazil at 2130.

Ian Howes of Lowestoft used a TV antenna with his R209 Mk. 2 receiver to log the following very interesting stations:

4500 Urumchi, China with music at 0010.

4665 Pathet Lao, Laos in Laotion at 1545.

4790 R. Ondas Portenas Venezuela at 0010.

4800 AIR, Hyderabad, India at 1545.

4840 AIR, Bombay, India at 1510.

4965 R. Santa Fe, Colombia news at 0000.

9680 VLH/R9 ABC, Melbourne, Australia at 1100.

11875 R. Dif. Nacional, Nicaragua at 0000.

11880 R. Splendid, Argentina in Spanish at 0045.

Peter Herman of New South Wales, Australia has an equipment line-up which includes a Trio 9R-59DS receiver, a 9MHz. dipole, a 15MHz. dipole, and a Hitachi SCT-115OR 3-band radio/cassette with a telescopic and 12 foot long-wire antenna. Peter's log included:

3322 R. Bongainville, N. Guin. at 1143.

4820 R. Gambia at 1800.

6145 AIR, India at 1540.

7100 R. Budapest, Hungary at 1630.

7235 All India Radio at 1500.

7285 R. Berlin International at 1730.

7290 R. Kuwait at 1630.

11920 R. Abidjan, Ivory Coast at 1830.

D. A. Hairon of St. Clement, Jersey has sent in another report using the usual equipment, this time he has heard:

9570 ABC, Australia in English at 1015.

9625 Israel B.C. in English at 2130.

11730 R. Nederland, Bonaire at 0525.

11760 R. Habana, Cuba, sign off at 0200.

11815 TWR, Bonaire in English at 0045.

11860 BBC, Ascension Is. relay at 0815.

11875 R. Nacional, Nicaragua, Spanish at 0130.

15160 R. Ankara in Arabic at 0530.

15532 R. Bangldesh in English at 1245.

17855 NHK, Japan in English at 0900.

21605 R. Kuwait in Arabic at 1100.

Alastair Nimmo is only eleven years old but this extract from his log, using a Meridian 10 transistor portable and 100 foot long-wire shows distinct promise:

5960 HCJB, Quito, Ecuador, English at 0830.

6040 VOA, Rhodes relay at 2100.

9515 R. Ankara in English at 2200.

9525 RSA, South Africa in English at 2215.

9530 AIR, Delhi in English at 1900.

9530 VOA, Monrovia relay, English at 2230.

In order to mention as many reporters as possible I will end this article with a few short extracts from the many received:

Richard Coyle, Glasgow, Lafayette KT340:

4965 R. Santa Fe, Colombia at 0550.

4970 R. Rumbos, Venezuela at 0345.

4990 R. Barquisimeto, Venezuela at 0400.

5038 R. Bangui, Ident. in French at 0500.

5075 R. Sutatenza, Colombia at 0100.

Adrian Pell, Wareham, Dorset:

12025 Voice of Vietnam, English at 2015.

15165 Danish Radio in Danish at 1400.

15295 TWR, Bonaire in Norwegian at 2145.

15370 VOA, Greenville in French at 2200.

Ian Newbold, Birmingham, R209 Mk. 2, 95 foot aerial:

3340 R. El Mar, Equador at 2000.

4550 R. Nacional, Colombia at 2005.

9735 NHK, Japan in Japanese at 2018.

15170 ELWA, Liberia in French at 2007.

Fred Wall, E.17, PR155, 20 foot long-wire:

3380 R. Malawi in English at 1800.

9510 R. Barquisimeto, Venezuela at 2350.

9520 R. New Zealand in English at 0900.

11930 Windward Is. B.S. cricket at 2120.

Philip Sokell, Barnsley, Romer 10, telescopic antenna:

6125 VOA noted at 1610.

6185 Radio Australia at 2045.

7215 Radio Cairo, Egypt at 0254.

7230 Radio Kiev, Ukraine at 0045.

Reports should arrive by the 15th of the month and be addressed to me at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.

WIT ZO



THE AMATEUR BANDS David Gibson, G3JDG

Frequencies in kHz • Times in GMT

In theory, it should have been a very good month for keen listeners since almost every log sent in was for bands which were at either end of the r.f. spectrum. Quite a few queries in the post bag, many asking questions which would require a text book, blackboard and a couple of years at evening classes to answer. Many people who have queries have a very simple way out—join the nearest radio club. You can ask a question and get an immediate answer. There is nearly always someone at the club who specialises in an area where your question is aimed at. If one person doesn't know the full answer, then it is virtually certain that someone else will.

Join the R.S.G.B., too. This organisation will give you all the help you need, and will also tell you the name and address of the secretary of your nearest radio club. The Society also publishes a number of "booklets" especially for the radio enthusiast and some are aimed specifically at the beginner.

Having joined a club, you will get a chance to visit other people's 'stations' and see gear which, at present, is only a number, like PR40 or R107. Better still, you can have a twiddle with the gear and make up your own mind as to whether a particular piece of equipment is very good or just mediocre.

So you don't hear any DX in spite of all those lovely logs you read in *Practical Wireless*? One way to latch on to some DX is to listen for a DX net which has European stations either in it or even running it. Listen for a good European station calling a DX station. Once you have found the right frequency it is highly probable that other stations will come up also calling the DX station. Eighty metres is a common hunting ground for this type of DX net.

Combined efforts of the **Ipswich** and **Colchester** clubs will result in the Anglian Mobile rally to be held at Ipswich on June 18. Talk-in stations on 160, 80, 4 and 2 metres. Colchester club also runs slow morse practice sessions. Anyone in the area might like to drop to Hon. Sec. a line at 26 Pondfield Road, Colchester, Essex. See you at the rally?

An appeal from Sam Elsdon for Amateur stations to use "standard" phonetics. I agree. Only way seems to be to make an accepted phonetic alphabet compulsory and written into all Amateur licences throughout the world. Sam has just finished off the CQ2 v.h.f. receiver which appeared in the September edition of Practical Wireless back in 1969. He runs a CR70A and PR40 plus a 310ft end-fed. A listen on eighty metres s.s.b. raised: DJ9NW, PJ4AQ, W5ILR/TF, XE1CV, 9H1D. Log for 20 metres reads: BY4AP, EA3AKE, FG7TC, IS1KLO, VE3MR, VK2AVA, VK30EL, VK50B, VO4HW, ZS5KY, 4X4DK, 7X2PK, 7X30RU.

Kevin Lamb has seen thirteen summers pass and resides at Ashford. Gear consists of a two-valve homebrew receiver, a.t.u. (also homebrew) and a 50ft end-fed. Goodies heard on topband include: HB9CM, OK1FT, OK1MAC, OL1APC, 4U1ITU.

Twenty metres using an R107 (that's cheating) raised: CX2XA, EA8GZ, EA8DI, HK5ASM, PY7AZQ, VK3ALL, VK4NB, VK4UC, VK5FH, VK6HE, VP9GK, ZL1HD, ZL3AH, ZL3AR, ZL3HA. Kevin asks about some four metre activity. This is a band which the Amateur could so easily lose unless there is a lot more activity. If the two metre addicts would come down, the G8 plus threes take the trouble to learn c.w., and the h.f. DX types go up another band, we might just save it—or is it worth the effort?

One hundred and thirty-two feet of wire stretching into the sunset but anchored to the aerial terminal of an R107 is a feature at a house in Melbourne Road, Chester. Mike Purcell loiters thereabouts and heard the following on 3.5MHz s.s.b.: CR4BS, CT1UN, HB9LQS, VE1QM, VP2LAT, VO1CU, WB0FFG/TF, W1AA, 4X4UF.

Another CQ2 v.h.f. receiver builder is **David** Lawley (Gravesend). He has heard a fair number of Amateur two-metre stations already but although he is only ten miles from the beacon station GB3VHF, he can't hear it at all. (V-e-r-y interesting.) **Ten** metres is the band which has brought a lot of stations in for David using the School Radio Society's B40 plus 500ft long, long, long wire. Log reads: JA8GWA, JRIINC, OD5HI, PZ1DV, VE3DOR, VE7HC, VU2JM, K1BCD, hoards of W stations, ZC4BJ, 4X4GH, 4X4HK, 9X5MS all mostly on s.s.b.

Down on topband, David reports signals from: E19J, GM3YOR, GW3UCB, GW3UPK, GW4AHN/A, HB9CM, HB9NL, OL1AOH, OL5ANJ, PA0PN. These were all on c.w. but PA0PN is on most Sunday mornings on s.s.b.

Eighty-three W stations start off the eight metre log of Richard Coyle (Glasgow). Additional appearances from: VE1ANZ, VE2WA, VE3GCS, VE3VE, ZL3GS, 6W8DY

Fifty-nine W stations in the ten-metre log from an unknown listener at Henrietta Street, Girvan in Ayrshire. OK, don't sign your name, but I know you've got a CR70A and a ten-metre dipole. Incidentally, it is surprising how many logs don't get in because no receiver is mentioned, or the aerial hasn't been divulged or the whole list is not in alphabetical order. (Hint! hint!!)

John Guy (Wimborne), B40C, 70ft end-fed, 28MHz, mode unspecified: CR4BS, CR611, ET3USA, FL8MM, HC2GG/P, KP4DHD, VS6DO, YV1AMX, YV3SZ, 7Q7BC, 8R1G.

More news from John Stevenson (Woking) about his d.c. (direct conversion) receiver. A number of mods have given a large increase in sensitivity. Log for 14MHz s.s.b. reads: CTIBZ, HR1RS, HV3SI, KP4BSA, PY1CAD, PY2YRS, PY4AS, VK2NN, VK2RV, VQ9R, ZL1BE, 3A2CP, 4X4BL, 4X4NJ, 4Z4TV, 9H1CZ.

Happenings for the merry month of May include: 6-7, 432MHz contest; 21, 2 metres contest. June: 3-4, National Field Day; 10-11, 4 metre contest.

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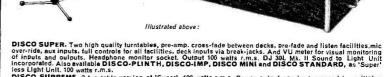
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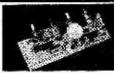
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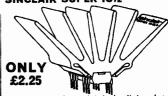
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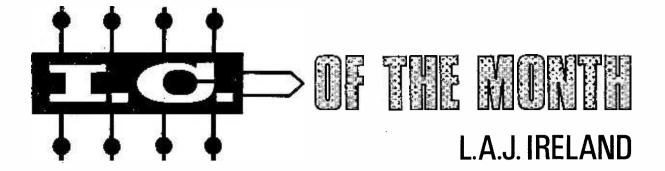
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Number 30

RCA CA3090Q Stereo Decoder

N AUGUST 1970 this column reported on the introduction of a stereo decoder i.c., the Motorola type MC1303. In the interval this unit has proved very successful and certain commercial concerns are marketing decoder units based on this device. However, technology does not stand still; it was recognised that the elimination of the set of coils required in a standard stereo decoder would facilitate the assembly of units on a commercial basis.

Successful decoding demands the selection of the 19kHz pilot tone from the audio signal developed by the discriminator stage of the f.m. tuner, followed by the reconstruction of the 38kHz carrier by frequency doubling without degrading the phase relationship with the pilot tone. Replacement of the carrier into the 38kHz sideband permits demodulation of the "difference" signal; the audio components of the discriminator output provide the "sum" signal.

The final stage in decoding is the reconstruction of the separate right and left channels from the "sum" and "difference" signals supplied. To reduce noise in the subcarrier (38kHz) regeneration process, it is desirable that sharply tuned circuits be used to reduce the bandwidth about 19kHz when determining the pilot tone. This however, is in conflict with the requirement of phase coherence, since then slight detuning can introduce significant phase shift. Remembering that a channel separation of at least 20dB is the minimum acceptable in a stereo decoder and that less than 10° of phase shift in each of the three inductors alone can cause a reduction in channel separation approaching this figure, the care necessary in designing and aligning the coils can be estimated, even with a sound basic system such as the MC1303.

At this point it may be recalled that the introduc-

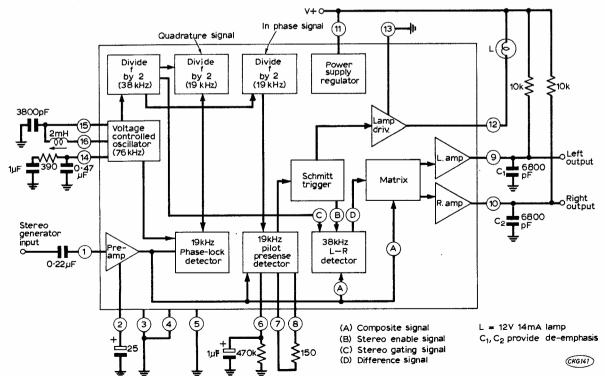
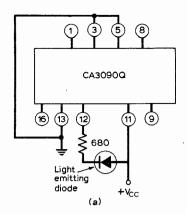
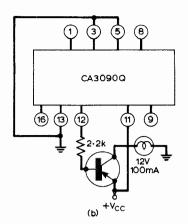


Fig. 1 : The R.C.A. type CA3090Q phase-locked loop i.c. stereo decoder block diagram and applications circuit. In domestic tuner systems the f.m. discriminator is applied across the 0·22 μF capacitor.





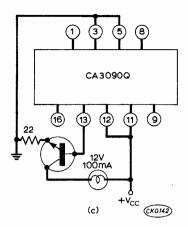


Fig. 2: Various methods of light indication (a) using a light emitting diode (b) PNP lamp driver (c) NPN lamp driver.

tion of monolithic silicon technology has opened an alternative route to the operation of tuned circuits, capable of eliminating to a great extent the requirements for precision LC circuits. It is, of course, the phase-locked loop circuit, first dealt with in "I.C. of the Month" in July 1971. Here, a local oscillator, whose operating frequency can be shifted by a limited amount by the application of an external voltage bias, drives a phase comparison circuit. The other input to the comparator is the external reference signal. The comparator output, a measure of frequency and phase separation of the local and the reference signal, feeds back to the local oscillator as an error voltage, correcting the local oscillator frequency.

The phase-locked loop i.c. considered last July was the Signetics NE561B system, suitable as an a.m./f.m. demodulator. This month exactly the same circuit concept is applied to the stereo decoder problem in the new R.C.A. Type CA3090Q. It is perhaps unnecessary to point out that by a phase-lock method, the problem of phase-shift degradation of channel separation is clearly avoided, and the need for design compromise between subcarrier circuit bandwidth and signal/noise ratio circumvented.

In the CA3090Q the loop local oscillator operates at a centre frequency of 76kHz and is followed by bistable frequency dividers, producing the 38kHz subcarrier at first division and 19kHz, the pilot tone frequency, at the second division. The frequency and phase comparison operation is carried out at 19kHz by reference to the received pilot tone, and the d.c. error voltage fed back to the local oscillator to maintain phase lock. Unlike the Signetics unit, the voltage controlled local oscillator of the CA3090Q is an LC oscillator, but the coil for this circuit is the only inductor required in the whole stereo decoder system, and even then alignment is non-critical. In point of fact, a 4kHz shift in local oscillator freerunning frequency requires a correction voltage representing some 10° total subcarrier phase shift; the 40dB. channel separation figure then achieved is highly satisfactory, Fig. 1.

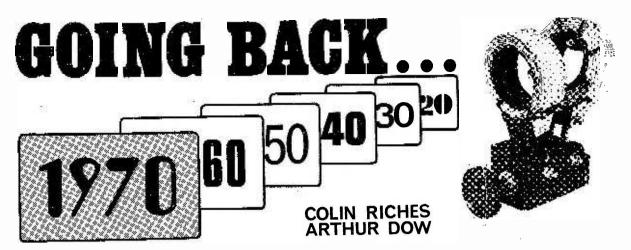
Signals at 19kHz are derived from the local oscillator by division independently for the phase comparator already mentioned and for a stereo signal presence detector circuit. When the pilot tone exceeds a preset value, corresponding to an f.m. detector output (i.e. the "composite" stereo signal comprising a.f. "sum" signal, 19kHz pilot tone and 38kHz suppressed carrier "difference" signal) of 40

mV, a Schmitt trigger operates, switching the i.c. into stereo mode, with demodulation of the 38kHz signal and matrixing for channel separation, together with a suitable stereo reception signal. External indication of stereo operation is therefore directly available in a dial light, with the attractive possibility that this may be an l.e.d. solid state indicator, rather than a mere bulb, Fig. 2. A manual override to the automatic mono/stereo switching operation may be incorporated; the control voltage to secure this function is applied at pin 4 (shown earthed in Fig. 1).

The circuit incorporates an internal voltage regulator, a feature widely accepted in complex function i.c.'s since it eliminates the need for external decoupling of power supply lines, a decoupling which could well be ineffective anyway due to internal "crosstalk" on the chip. Further, in a complex circuit, the availability of pins for external connections is often a limiting factor, and economically more significant in production than a few extra transistors on the chip, which require consideration once and for all only at the design stage. The effectiveness of the internal regulation is indicated by the capability of the unit to function over a power supply voltage range of 10 to 16 volts.

Design of the actual decoder can be based on Fig. 1, which shows the actual circuit for a phase locked loop automatic stereo decoder with function indicator lamp. The coil is a 2mH unit, and may be wound on an adjustable pot core. The complexity of the unit can be judged by the fact that it has 128 transistors! Clearly an itemised analysis of circuit function would not find a place in this note, so Fig. 1 also provides a block diagram to assist the constructor who wishes to study the system more fully. For the majority of readers, though, it will be sufficient that in this i.c. there is available a first-rate stereo decoder which is simple to apply even if highly sophisticated in design.

The unit is supplied in a 16-pin quad-in-line plastic package, i.e. differing from the d.i.p. package in that alternate pins are displaced to provide four lines each of four pins. This facilitates the design and construction of suitable p.c. boards; it is recommended that an earth strip be left down the centre of the i.c., screening contacts on one side of the i.c. from those on the other. However, no difficulty should be experienced, even by the relative newcomer to construction, in achieving success with this unit.



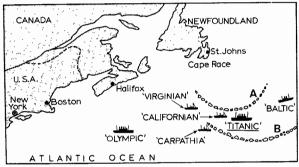
TITANIC DISASTER - PART 2

Mr. Cyril Evans, examined by the Solicitor-General at the Court of Enquiry on the disaster, stated that he was the sole Marconi operator in the Californian. At 5.35 p.m. New York time, or 7.30 ship's time, he received a message from the steamship Antillian that an hour before she had seen three large icebergs to the southward. A little later he heard from the Titanic and offered her the report about the ice, and she replied, "All right, I heard the same thing from the Antillian". At 9.5 New York time, or 11 o'clock ship's time, the captain directed him to tell the Titanic that the Californian was stopped and surrounded by ice. He sent the message to the Titanic and got the reply, "Keep out." That was because the Titanic was at that moment in communication with Cape Race, and his message had caused an interruption. The *Titanic*, however, must have heard what he had said about the ice, because his signals were much stronger than the Cape Race signals. He next heard the Titanic say to Cape Race, "Sorry, please repeat". The messages from the Titanic to Cape Race were private messages from the passengers.

Mr. John Durrant, Marconi operator of the *Mount Temple*, was another witness. In reply to the Solicitor-General, he stated that the range of his wireless installation was 150 miles by day and 200 miles by night. On the evening of Saturday, April 13th—the day before the foundering of the *Titanic*—he got



H. T. Cottam, wireless operator on board the "Carpathia".



Map showing the approximate positions of the Titanic and other ships.

an official message from the captain of the *Mount Temple* that ice had been seen. This was the only message he received in regard to ice before the wreck.

The witness then proceeded to give from his log-book the various calls he heard sent by the *Titanic* and the replies to them by ships which they reached.

The first thing he heard of the Titanic was at 11 minutes past 12 o'clock (ship's time) on Sunday night, when he got the message "C.Q.D." from the *Titanic*, giving her position, and adding, "Come at once. Struck berg. Advise captain." He told his captain at once. After the lapse of ten minutes he had the entry, "Titanic still calling C.Q.D.", that she was asked by the Carpathia what was wrong, and replied, "Struck iceberg. Come to our position," which was given. At 12.26 a.m. he made the entry-"Titanic still calling C.Q.D." At this time the Mount Temple had altered her course, and was speeding to the assistance of the *Titanic*. This had been done about 15 minutes after getting the first signal. At 12.34 he heard the Frankfurt answering the Titanic and the Titanic giving her position to the Frankfurt. The *Titanic* asked, "Are you coming to our assistance?" The *Frankfurt* said, "What is the matter with you?" and the Titanic answered, "Have struck an iceberg. Sinking. Come to our help. Tell captain.' The Frankfurt then said, "O.K. Will tell bridge at once", and the Titanic replied, "O.K. Yes. Quick." At 12.42 he heard the Titanic calling "S.O.S."

At a quarter to 1 o'clock he heard the *Titanic* sending out both calls. She then got into touch with the *Caronia*, and next with the *Virginian*.

The Solicitor-General then asked if Mr. Durrant had broken in and talked to the *Titanic* would he

have interrupted her messages to other ships? Yes, I never said a word after I got her position. The first rule in wireless telegraphy is "Never Interfere".

The witness, continuing the narrative from his log-book, said the *Titanic* called the *Olympic* at 12.43 a.m. The *Olympic* replied at 1.06 a.m. and got the message, "Get your boats ready. Going down fast by the head." At 1.11 the *Frankfurt* sent a message to the *Titanic*, "Our captain will go for you". At 1.13 he heard the *Titanic* working the *Baltic*.

The witness said the *Titanic* answered the *Olympic*, "We are putting the women off in the boats". At 1.29 the Titanic sent out a general call. "C.Q.D. Engine-room flooded". The Titanic also informed the Olympic that the sea was clear and calm. At 1.31 he heard the Frankfurt say to the Titanic, "Are there any boats round you already?" and to this the Titanic made no reply. At 1.33 he heard the Olympic send a message to the Titanic asking whether the Titanic was steering south to meet the Olympic and the reply of the Titanic was simply the code word for "Received". That was the last message that he heard from the Titanic. The messages from the Titanic did not get fainter towards the end. When the messages ceased, he thought the flooding of the engine-room had put the wireless out of condition. Most ships, including his own. carried storage batteries for use when power could not be obtained from the dynamos, and the wireless apparatus could be changed from the dynamos to the storage batteries in a minute; but the range of a wireless using storage batteries would be less than that of a wireless using dynamos.

At 1.41 a.m. he heard the *Frankfurt* and the Russian ship, the *Birma*, calling the *Titanic* and there was no reply. At 1.56 the *Olympic*, the *Frankfurt*, and the *Baltic* called, and again there was no



Mr. John Durrant, operator on the "Mount Temple".

answer from the Titanic. At 2.11 the Birma informed the Frankfurt that she was 70 miles from the Titanic. At 2.36 he made the entry, "All quiet now. The *Titanic* has not spoken since 1.33." At 3.11 he heard the Carpathia say, "If you are there, we are firing rockets." At 3.26 the Carpathia again called the Titanic. At 3.44 the Birma told the Frankfurt that he thought he heard the Titanic and calling her, said, "Steaming full speed to you. Shall arrive 6 in the morning. Hope you are safe. We are only 50 miles away." At 3.46 he heard the Carpathia calling again. At 4.40 he made the entry. "All quiet. We are stopped away. Pack ice." At 5.11 the Californian called "C.Q.", and he answered telling her that the Titanic had struck an iceberg and sunk, and he gave her the position. At 5.26 he heard the Californian speaking to the Frankfurt, and she replied to the same effect. His last entry was, "8 a.m. Heard from Carpathia that she had rescued 20 boatloads".

P.M.G. Comments

The Right Honourable Herbert Samuel, M.P., Postmaster-General, referring to the disaster at the dinner of the London Chamber of Commerce on April 18th, 1912, said:

"Those who had been saved had been saved through one man, Mr. Marconi, whose wonderful invention was proving not only of infinite social and commercial value, but of the highest humanitarian values as well." He had seen it stated that in the United States of America the efficiency of the wireless telegraphy service had been impaired by lack of regulation. He did not know whether that was well founded or not, but as Postmaster-General he could assure them that such disturbance was impossible here. Parliament had given the Postmaster-General a complete control over the use of wireless telegraphy, and no one could operate or establish a station without the Postmaster-General's licence, which was only very sparingly given, and for purposes of experiment and research and under such conditions which precluded disturbance of commercial or humanitarian messages. Round the coast, in charge of his department, there was a girdle of wireless stations which were in constant communication with the telegraphic services of the country and with the life-saving stations. No fewer than 400 liners had been equipped with wireless apparatus, including a certain number of cargo vessels. All the operators on these ships were required to hold a Post Office Certificate of Efficiency, and to answer immediately any signals of distress, and under conditions which, as far as possible. precluded interference with one another.

David Sarnoff

David Sarnoff was on duty at the Marconi Wireless Telegraph Co. of America station at Saisconset on Nantucket Island. On the night of April 14, 1912, he picked up the messages that told of *Titanic's* distress signals and he stayed on duty continuously for 72 hours so that he could relay messages from the rescue ship to the rest of the world.

Brig. General David Sarnoff, born in a small village near Minsk in Russia, in 1891, and former Chairman of the Board of R.C.A. passed away on December 12th, 1971.

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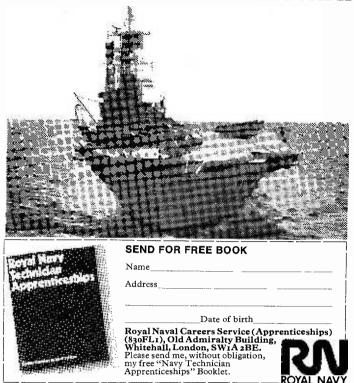
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250V P.C. moun 3½p. 1μF, 4p. 0·1 1·5μF, 20p. 2·2μF Miniature Fixed 63V dc wkg. 1·8p 12pF, 15pF, 18 120pF, 150pF, 18	ting: 0.01µF, 0.5µF, 0.22µF, 5p. F, 24p. Ceramic Plate. CoF, 2.2pF, 2.7pF, 2.2pF,	015μF, 0·022μ 0·33μF, 6½p. 0· 333 Series, 3p ε , 3·3pF, 3·9pF, 33pF, 39pF,	F, 3p. 0.033 47μF, 8½p. 0 ach. 4.7pF, 5.6p	68μF, 1 F. 6.8pl	.1p. 1·0μF F. 8·2pF.	, 13p. 10pF.	OC28 1-9 10 plus	32p 30p
ELECTROLYTIC (μF/V) 10/2·5, 40 400/4, 6·4/6·4, 2i 125/10, 200/10, 2 25/25, 50/25, 80/2 20/64, 32/64.	CAPACITORS /2-5, 80/2-5, 160/ 5/6-4, 50/6-4, 10/ 2-5/16, 10/16, 20	MULLARD C42 2-3, 320/2-5, 50 0/6-4, 200/6-4, /16, 40/16, 80/1	0/2·5, 8/4, 3 320/6·4, 4/10 6, 125/16, 1	6/25, 6	4, 125/4, 1 , 32/10, 1-4/25, 12	64/10 5/25	OC35 1-9 10 plus	33p 28p
MULLARD C437 100/40, 160/25, 2 400/16, 640/10, 2000/4, 1000/10, 1600/10, 2500/6-4	50/16, 400/10, 64 1250/4, 1000/6·4, 1600/6·4, 2500/2	, 1600/2 5, 12p	. 160/64, 25	0/40, 40	00/2.5, 64	10/16.	2N 3055 1-9 10 plus	49p 46p
RESISTORS 1 watt 10% carbo 2 watt 10% carbo range 2.2 ohms to triple rated 1-1	on 1 10 megohms typ -}, tin oxide	p each 7400 p each 51, 5 e TR5 7442 ±2% ★ S	GAIN TTL's, 01, 02, 03, 3, 54, 70, 72, 73, uper Low P.	74, 76, 8	20, 30. 40 35p 36.	each	2N 2926 G 1-9 10 plus	10p 8p
range 10 ohms to MINIATURE NE HB1725 1.9mA 2 SB1725 0.6mA o	1 meg. ON LAMPS 51 30v.	p each 709c 709c 741c 741c 747c	TO99 . DIL . TO99 . DIL .		::	28p 32p 38p 36p 42p	MP8111 1-9 10 plus	32p 28p





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Normal Price £65 00 aluminium trims. **REALISTIC 30 WATT STERFO** A superb hi-fi amplifier with all the fea-tures you've ever wanted – for under £46-00. Saving over £10-00 on the normal

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and noise: phono - 50db; tuner/aux - 65db. How's that for a specification!
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OLSON RA-310 AM/FM/MPX STEREO TUNER This ROC Tuner is especially de-

signed to match the Dison AM-395 Stereo Amplifier. In price and value, as well as it's good look ing design! But of course it's also deal for use with any other amplifier. The RA-310 costs £10-00 less than

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February 1972

the normal retail value, and yet it is a highly sophisticated unit. Incorpo-rating the latest solid state techniques. Operation is drift free for sup-50.00 reme station-holding capability. You can connect this Tuner to a stereo amplifier, to a tape deck or a tape recorder. And of course it covers all the stations in the AM and FM bands. FM: 87-108 MHz; AM: 525-1605 kHz.

FM Sensitivites: FM. 3µV: AM. 250µV. Stereo separation 30dB at 1kHz. image rejection 60dB. Size: 11 + "wide, 4" high. 7½" deep.

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Here's fabulous, exciting value in miniature! This high quality stereo amplifier measures only 8" wide × 3" high × 5%" deep. And yet it has sepa rate ganged volume, balance and tone con-

trols. Plus speaker in/out, mono/sterco, phono uner and power on/off slide switches. The ends are oiled walnut, with match £12.45 ing enamelled metal top. The front panel is satin aluminium and walnut-brown enamel. Frequency response is 50 to 10,000 Hz \pm 3dB. Output 3 watts r.m.s. per channel into 8 ohms. Inputs are 100mV for both phono and tuner.

REALISTIC TM-100 STEREO TUNER Here's another unit that gives

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Designed specifically to match
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40-WATT STEREO
AMPLIFIER
An ideal unit for your new

stereo separate system. It is more than £10.00 below the normal retail price! Making the AM-395 one of Britain's best hi-fi

buys. It takes in signals from magnetic or ceramic pick-ups, tuners (see Olson RA-310) and tape decks. And it's got outtuners (see Dison RA-310) and tape decks. And it's got out-by the property of the second of the headpholes. There are separate bass and treble controls, separate Left and Right' channel volume controls. And a loudness switch for boosting the bass and reble notes when listening at low output levels. Frequency response: 20-20.00 Mt ± 388. Output; 20 watts r.m.s. per channel into 8 ohms. 100 mts. 'angular behood 30 mt RAAA, crystal phon 0 100 mt; tape 160 mt; tuner 160 mt. Size 11 + "wide. 4" high, 7½" deep. The specification redat well—scords area better.

specification reads well - sounds even better!

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STEREO CARTRIDGE TAPE DECK MODEL RP-1000ST

The popular Lear-Jet type recording unit is the heart of the fantastic RP-1000ST, which has full record and play-back facilities. Automatic track change with manual override Press to start button. Stereo headohone and Left and ride. Press to start button. Stereo headphone and Left and Right Mic sockets on front panel. Dual recording level meter, and Left and Right volume controls. Built-in preamp. Tape speed 3½ ips (9.5 cms). Frequency response: playback 30-10.000 Hz; recording/playback 30-8.000 Hz. Line output: fully variable 0-500mW. The RP-1000ST incorporates all the features you'd expect in a top quality Cartridge Tape Deck. Size 16* wide. 4* high. 9* deep. Cabinet in walnut, lackuding connecting leads



-TRACK HOME STEREO CARTRIDGE PLAYER 🂽 MODEL ET

MODEL ET With this unit, you can play any standard 8-track cartridge on the market — at a fraction of the normal retail value! It gives you a total of 5 watts ol power, to feed into two 8-ohm speakers. The fraquency response is 50 to 10,000 Hz. giving you a fine tonal quality that can't be bettered at anything near this price. The E1 has separate tone. balance and volume controls, giving you complete free-dom to select the sound you want to hear. Tape speed is $3\frac{2}{3}$ ips, and wow and flutter are both less than 0.3%. Size: $11\frac{1}{2}$ wide, 5,5 high, 11 deep.



R.446 3-WAY MATCHEO SPEAKERS

These will do justice to your amplifier - and to your pocket. Atonly £17-80 a pair, they are real value-for-money. Each cabinet is heavily legacd and teak finished. They handle 16 watts rms (8 watts rms each). Each loudspeaker contains a large dual cone base unit, plus a separate tweeter. Frequency range: 40 to 19.000 Hz. Size 14" high. 9" wide, 61" deep.

OLSON AM-357 4-WATT STEREO AMPLIFIER RHER'S marvellous value for someone just starting to set themselves up in audio! At only £9-00 you get a fine amplifier in a scratch resistant metal cabinat, with a smart brushed aluminium front panel. It incorporates separate tone and volume controls for each channel. Inputs are provided for turntable (ceramic car-tridge), tuner and tape deck or recorder. Frequency response: 70-20.000 Hz ± 3dB. Output: 2 watts r.m.s. per channel into 8 ohms. Inputs: phono 80mV; tuner/aux 80mV. Size 8" wide, $2\frac{\pi}{8}$ " high, $4\frac{\pi}{2}$ " wide.

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Headphone socket on front panel for easy access, Single tuning kaob for FM and AM.
Built-in aerials, facilities for external FM aerial. Illuminated dial. Steree broadcast indicator light. Each speaker has 63° bass and 3" treble units, Dulput: 9 watts r.m.s. per channel into 8 ohms. Frequency response: 30 to 20.000 Hz. FM: frequency range 80-108 MHz, sensitivity 2 5µV, stereo separation 30d8. image rejection 40db.

AM: frequency range 530-1605 kHz, sensitivity 100µV.
Sizes: speakers 8, wide, 12', high and 93', deep; freceiver 6* wide, 12' high and 93', deep. And their sophisticated appearance matches the excellence of the specification.

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giving you the ideal frequency response to hi-fi, natural or mood music listening. Its beautiful, heavy, oiled walnut cabinet

incorporates two separate speaker units an 8" woofer, and a 5 mid-range with 2" concentric tweeter. Power handling capacity: 25 watts r.m.s into 8 ohms. Overall frequency res' ponse: 35-20,000 Hz. Cabinet size: $10\frac{1}{2}$ \times $7\frac{1}{2}$ \times $8\frac{3}{4}$. Exactly right for matching the most modern decor.



PALACE AM/FM/MPX STEREO TUNER AMPLIFIER SSA-16

PALIACE AMI/FM/MMY-STEREO TUNER AMPLIFIER SSA-16
This is one of the lowest priced stero tuner amplifiers on the market. It covers the full range of both AM and, FM broadcast frequencies. And when you're switched to FM. an indicator lights up when a starea signal is received —that's the time to switch to 'Stereo'! The SSA-16 has all the facilities you'd expect to find on tuners octing twice as much — separale volume here. yme. bass, treble, balance and tuning controls, Selector switch for tape; phono, AM. FM, stereo, Jack socket on front panel for stereo haddhones. Frequency range: FM 88-108 MHz. 535-1605 kHz. Frequency response: 50-10,000 Hz. ± 348. Power output: 4 watts total music power into two 8 ohm speakers. Size: 15" wide, 4½" high, 8" deep.

ROC 7-WATT STEREO AMPLIFIER CHASSIS SK-317 This exclusive R O C Stereo Chassisis completely self-contained, and it costs £2.25 less than the normal re tail value! T SK-317 is a really The

compact unit measuring only 5½" wide, 1½" high and 6½" deep. It contains its own mains power supply, and has a ganged tone control and separate volume controls for each channel. Specification: frequency response 40-17,000 Hz ± 3dB; output 3-5 watts music power per channel into 8 ohms; input, phono, 600mV; signal-to-noise i ratio better than 45dB.



DISON AM-372 16-WATT, STEREO AMPLIFIER A: Here's a really good amplifier at a really down-to-earth price – nearly £7 less than the normal retail value! Just look at what the AM-372 will do you - reproduce signals from ceramic or crystal cartridges, AM and FM tuners, and tape recorders. And it gives you outputs for two sets of speakers, headphones and tape recorders. Frequency response is 30 to 20,000 Hz ± 3dB. Dutput B watts r.m.s. per channel music power into 8 hm speakers. Phono input 200mV. Tuner input 200mV. Size: 121" wide, 31" high, 71" deep.



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mpedance 8 ohms per channel. ROC PRICE £2-95

EAGLE SE-30 STEREO HEADPHONE

This model is for the more discriminating listener. For a start the frequency range ex-tends from 30 to 16,000 Hz. And you can adjust the volume of each earpiece inde-pendently. There's also a mono/stereo switch. For maximum comfort, the ear cushions as covered in soft leathers, ROC PRICE £7.05



TEC HR-007

problem. Separate volume and tuning controls with easy-to-use knobs. Frequency range is

535 to 1605 kHz medium wave band. Maximum output is 300 mW. Normal Price £9:45 RDC PRICE £5:86



8-TRACK CAR

Orive to the sound of music this fabulous 8-Track Cartridge Player. It gives you superb tone and power to fill the car with stereo sound. Ideal for use with R.151 or R.152 speakers. Complete with all mounting accessories. For negative earth electrical systems only, Output: 2-5 wats per channel. Frequency range: 70-10,000 Hz. Wow and flutter: less than 0-3%. Tape speed: 3-5 cm/sec. Channel selector: automatic with manual over-ride. Mounting dimensions: 532"×52"× 2H" ROC PRICE £27-20

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that could only reproduce ceramic or crystal pick-up cartridges, can ceramic of crystal pick-up cartridges, can accept signals from moving-magnet cartridges 1 The R.307 steps up signals from between 5-20mV to 200-800 mV. Input: 5-20mV. Equalisation: RIAA. Dutput: 200-800mV flat. Frequency range: 20-2000 Hz. Dimensions: 32, 72, 82, 47. Supply: 140 VAC. ROC PRICE £4-92

15-FOOT STEREO HEADPHONE
EXTENSION CORO R.362
Fitted with heavy duty
3-circuit stereo plug at one end and a matching stereo socket at the other. ROC PRICE £1:30

STEREO HEADPHONE "Y" ADAPTOR R.361 Enables you to use two sets of stereo headphones from a single socket. Fitted with male plug and two female sockets. ROC PRICE £1-30

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If you want easy, fingertip control headphones and loudspeakers, here's the idea solution to the problem. All you do is connect it to your speakers and amplifier, plug in you headphones-and you're ready to take over! At the flick of a slide switch, you can have headphones alone, or speakers alone, or both together. Input: suitable for use with amplifiers rated up to 20 watts. Size: 252 × 372 × 172 ...



BOC PRICE CT-50

R.151 STERFO

Smart black, tough, plastic cases, each containing a high flux 110mm diameter speaker unit. Just what you need to go with the CS.8 Certridge Player or any other car stereo system. Fitted with over three yards of connecting cable. Dimensions: 644"×545"×345". Impedance: 8 ohms per speaker. Rating: 5 watts max per speaker. ROC PRICE £3-72

R.152 STERFO CAR



FAGLE LC 05 STEREO MAGNETIC CARTRIDGE For fabulous reproduc-

For fabulous reproduc-tion at a very low price, you'll field it hard to beat. 0-7 mil diamond stylus. Dut-put: 6mV per channel. Frequency range; 30-18,000 Mz. Channel balance: ±1-5dB. Channel separation: 20dB. Recommended stylus pressure: 2-4 grams. Compliance: 9×10-6 cm/dyne. ROC PRICE E4-75

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A-3000 36-WATT SOLID STATE STEREO AMPLIFIER The A-3000 looks as good as it sounds! Giving you a big performance this superb audio amplifier has a full range of facilities on the front and rear panels. On the front -all the controls you're ever likely to need plus a headphone socket. On the rear signal inputs, speaker outputs and a line fuse for circuit protection. Specifications: 18 watts rms per channel

into 8 ohms. Frequency response 20-35,000 Hz (± '2db) Inputs Magnetic. Ceramic, Tuner, Tape, Aux. Tape Play. Size: 345mm × 300mm × 130mm. Normal Price £30.70. ROC PRICE £28.00



R-200 20-WATT AM/FM/MPX STEREO TUNER AMPLIFIER What more could a hi-fi enthusiast want! The R-200 gives you top quality reproduction of both AM and FM programmes, including all the stereo broadcasts now available on FM. And you have built-in facilities for recording your favourite programmes on an external tage recorder. The front page is carefully designed, with the letest slider controls for bass, treble, volume and balance. Alongside the dial are a meter for accurate tuning and a stereo indicator lamp that automatically lights up when you're tuned in to a stereo signal. Dozens of other brilliant facilities including main and remote speaker terminals

speaker terminals. Specifications: 10 watts rms per channel into 8 ohms. Frequency response: 25-40,000 Hz (\pm 2db) Inputs: Magnetic, Ceramic, Tape, Aux. Tape Play. Siza: 398mm \times 287mm \times 108mm. Normal Price. £50.00. ROC PRICE £42.00



CASSETTE DECK Precision engineered for trouble-free performance, this Stereo Cassette Deck has a fantastic range of facilities, making it a real value-for-money unit. Left and right level meters for recording, the latest slider controls for record level, switchable playback noise filter. aux/mic switch, mic input sockets, piano-key controls for record, rewind, play, fast-forward, and stop/eject. Index counter with reset button. Specifications: Frequency response: 35-12,000 Hz. Wow and flutter less than 0-25%. Inputs: Mike, Aux. Din Socket. Size: 345mm× 300mm×100mm, Normal Price £39-50 ROC PRICE £34-00

SAQ-501 50-WATT SOLID STATE STEREO AMPLIFIER A really powerful unit with all the facilities you need for home entertainment

inputs for magnetic cartridge. tape, radio tuner and auxiliary Controls for bass, treble, balance and volume. Head-phone socket on the front panel for easy access. Loudness switch. Rumble and scratch filters. Specifications: 25 watts rms

per channel into 8 phms. Inputs Magnetic, Tuner, Tape/Aux. Tape play. Frequency response 20-20.000 Hz (± 1db). Size: 333mm × 102mm × 285mm. Normal Price £33-60. ROC PRICE £26-40



A - 5000 60 - WATT

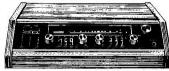
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With the A-5000, you're in the big-sound class -30 watts rms per channel into 8 ohms! The circuit is all-silicon-transistor, giving you top quality sound and a mere 0-2% distortion at 25 wetts output. And optimum stereo input balance is derived from the use of an IC (integrated circuit). There's no need to worry about overload or short-circuiting the output — the A-5000 has built-in protection, Specifications: 30 watts rms per channel into 8 ohms. Frequency response 15-40,000 Hz (± 2db) Inputs Magnetic, Tuner, Tape/Aux, Tape Play. Normal Price £43-40. ROC PRICE £34-00



R-150 12-WATT AM/FM/MPX STEREO TUNER AMPLIFIER You couldn't get batter value for money in Stereo Tuner Amplifiers anywhere! Just look at in Stereo Luner Amplifiers anywhere! Just look at all the facilities the R-150 gives you - bass, treble, balance, volume, switchable AFC for drift-less reception on FM, socket for hasdphones on the from panel. A tuning meter, Stereo indicator, Tape output, so that you can record your favourite programmes. To name but a few. AM section covers the medium waveband -535-1605 kHz, and the FM band 88-108 MHz.

303-100 KnZ. Bnd the FM band 88-108 MHZ. Specifications: 6 watts rms per channel into 8 phms. Frequency response: 40-20.000 Hz (± 2db) Inputs: Magnetic, Ceremic, Aux. Size: 107mm × 385mm × 263mm, Normal Price £38-30, ROC PRICE £29-90



MR-15 AM/FM/MPX STEREO TUNER AMPLIFIER Here's a baautifully styled AM/FM Stereo Tuner Amplifier. Featuring FET (Field Effect Transistor) front end FM tuner. Featuring FEI (Field Effect Transistor) front end FM tuner, and dual-channel IC equalizer for perfect balance, the MR-15 incorporates professional style vertical sliding controls for bass and trable. All the input/output facilities you eeed. Covers FM 88-108 MHz. AM 535-1605 kHz. Output 16 watts rms per channel into 8 others. Inputs Magnetic. Tape. Aux. Tape Play. Size: 467mm×458mm×130mm. Normal Price £67-60. ROC PRICE £54-00

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ECC85 40p EM84 35p PL84 35p TT21 22.78 ECC86 50p EM87 55p PL500 73p U25 78	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	805 £8·00p JP9/7D 807 50p £37·50p 813 £3·75p K301 £5·00p
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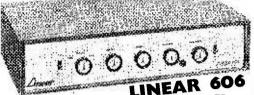
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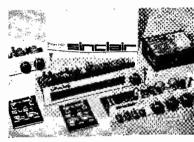
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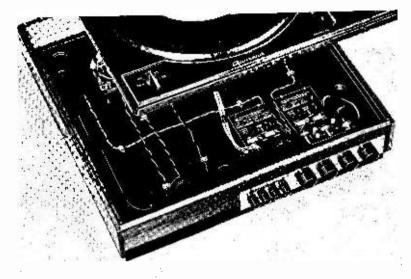
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Typical Project 60 applications

System	The Units to use	together with	Cost of Units
Simple battery record player	Z.30	Crystal P.U., IZV battery volume control	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U. volume control etc.	£9.45
20 + 20 W. stereo amplifier for most needs	2 x Z.30s, Stereo 60, PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
20 + 20 W. stereo amplifier with high performance spkrs.	2 x Z.30s, Stereo 60, PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc.	£26.90
40 + 40 W. R.M.S. de-luxe stereo amplifier	2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43
F.M. Stereo Tuner (£25)	& A.F.U. Filter Unit (£5.98)	may be added as required.	

Project 60 Stereo F.M. Tuner

£25 Built and tested. Post free.



The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz. Capture ratio: 1.5dB. Sensitivity: 7μV for lock-in over full deviation. Squelch level: 20μV. Signal to noise ratio: >65dB. Audio frequency response: 10 Hz – 15 KHz (±1dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: 2μV. Cross talk: 40dB. Output voltage: 2 x 150mV R.M.S. Operating voltage: 25-30 VDC.

Indicators: Stereo on ; tuning. Size: 93 x 40 x 207mm.

Stereo 60 Pre-amp/control unit

£9.98 Built, tested and guaranteed.



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

SPECIFICATIONS—Input sensitivities: Radio — up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A curve ±1dB:20 to 25,000 Hz. Ceramic p.u. — up to 3mV: Aux — up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE + 12 to —12dB at 10 KHz: SASS + 12 to —12dB at 10 OHz. Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm.

A.F.U. High & Low Pass Filter Unit

£5.98 Built tested and guaranteed.



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages – rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V. Current – 3mA. H.F. cut-off (—3dB) variable from 28KHz to 5KHz. L.F. cut-off (—3dB) variable from 25Hz to 100Hz. Distortion at 1 KHz (35V. supply) 0.02% at rated output. Size: 66 x 40 x 90 mm.

Z.30 & Z.50 power amplifiers

Built, tested and guaranteed with circuits and instructions manual. 2.30 £4.48 2.50 £5.48



The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at 15w (8 Ω) and all lower outputs. Whether you

SPECIFICATIONS (Z.50 units are interchangeable with Z.30s in all applications).

Power Outputs
2.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts.

Z.50 40 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts. Frequency response: 30 to 300,000Hz±1dB

they are the same size and may be used with other units in the Project 60 range equally well. Distortion: 0.02% into 8 ohms.
Signal to noise ratio: better than 70dB unweighted.

use Z.30 or Z.50 amplifiers in your Project 60

system will depend on personal preference, but

Input sensitivity: 250mV into 100 Kohms (for 15w For speakers from 3 to 15 ohms impedance Size: 14 x 80 x 57 mm

Power Supply Units



Designed special for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 where a stabilised supply is essential.

PZ.5 30 volts unstabilised £4.98 PZ.6 35 volts stabilised £7.98 PZ.8 45 volts stabilised (less mains transformer) £7.98 PZ.8 mains transformer £5.98

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If within 3 months of purchasing Project 60 modules directly from us. you are dissatisfied with them. we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that this early within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.



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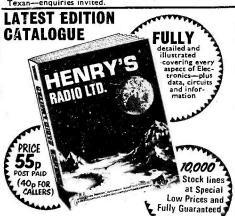


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