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rear, 2 tin front $(H)$. Comp with connecting leads. rear, ${ }^{2 \times 1 \text { in }}$ fron
Pair

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| $1 T 4$ | -16 | 30 Cl 7 | . 78 | EAF42 | . 50 | EM81 | . 88 | PCL84 | .84 | UBC41 | . 58 |
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| 20P4 | . 92 | DL96 | . 88 | EL84 | . 28 | PCF805 | . 81 | $\begin{aligned} & \text { U261 } \\ & \text { H301 } \end{aligned}$ | . 68 | ${ }_{\text {OC82 }}^{\text {OC8 }}$ | . 12 |
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|  | 300 m | 750 n | 1A | 1.5A | 3 A | 10A | 30 A |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{aligned} & \mathrm{sp}_{0} \\ & 0.05 \end{aligned}$ | ${ }_{0}^{\text {fp }}$ | ${ }_{0}^{\text {fp }}$ | A 0.14 | ${ }_{0}$ | ${ }_{0}^{28}$ |
|  | 0.04 | 006 | 005 | 0.13 | 0.16 | 0.23 | 0.75 |
|  | 0.05 | 009 | 0.08 | 0.14 | 0.20 | 0.24 | 1.00 |
|  | 0.08 | 0.13 | 0.07 | 0.20 | 0.27 | 0.37 | 1.25 |
|  | $0 \cdot 07$ | 0.16 | 0.10 | 0.28 | 0.34 | 0.45 | $1 \cdot 86$ |
|  | 0.10 | 0.17 | 0.13 | 0.25 | 0.37 | 0.55 | 2.00 |
|  | 0.11 | 0.25 | 0.15 | 0.30 | 0.48 | 0.68 | 2.50 |
| 1200 |  | 038 |  | 0.83 | 0.57 | 0.75 |  |
|  | triacs |  |  | LUCAS SILICON |  |  |  |
| 0 | 2 A | 6A | 10A | 35 amp, type. |  |  |  |
|  | T0-1 | T0-66 | то |  |  | $\begin{gathered} 00 \mathrm{p} . \mathrm{i} . \mathrm{i} \\ \mathrm{in} \cdot 10 \mathrm{p} \end{gathered}$ | $\begin{aligned} & \text { v. stud } \\ & \text { peach. } \end{aligned}$ |
|  | ep | fp | £p |  |  |  |  |
| 100 | 30 | 50 | 76 |  |  | Iacs |  |
| 200 | 50 | 60 | 90 | $\underset{\text { FOR }}{\text { FRI }}$ | ACB |  |  |
| 400 | 70 | 75 | 1-10 | BR1 | 00 (D | 32) 37 | each |

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EQVT $\begin{array}{ccc}\text { Pak No, } & \text { EQVT } \\ \text { T1 } & 8 & 2 \mathrm{G3713} \\ \text { T2 } & 8 & \text { OC71 } \\ & \text { D1374 } & 0 \mathrm{C} 75\end{array}$ $\begin{array}{cccc}\text { T2 } & 8 & \text { D1374 } & \text { OC75 } \\ \text { T3 } & 8 & \text { D1216 } & \text { OCS1D }\end{array}$ $\begin{array}{llll}\text { T3 } & 8 & \text { D1216 } & 0001 \mathrm{D} \\ \text { T4 } & 8 & 2 \mathrm{G3317} & \text { OC81 } \\ \text { T5 } & 8 & 3 \mathrm{G} 389 \mathrm{~T} & \text { OC42 }\end{array}$
 $\begin{array}{llll}\mathrm{T} & 8 & 2 \mathrm{G3} 34 \mathrm{BB} & \text { OC45 } \\ \mathrm{T} & 8 & 2 \mathrm{G} 378 & \text { OC78 }\end{array}$ $\begin{array}{cccc}\text { T9 } & 8 & 2 G 399 A & 2 N 1302 \\ \text { T10 } & 2 & 2 G 417 & \text { AF117 }\end{array}$ T10 y $2 G 417$ AF117
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V41 25 RF germ. trans. TO-1 OC45 NKT72
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| $\overline{0} 43$ | 25 |
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| Code | Powet | Tolerance | Range | Values | 1 to 9 | 10 to 99 | 100 up |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | in ohms |  |  | (see note below) |  |
| C | 1/20w | 6\% | 82-220K | E12 | 9 | 8 | 7 |
| 0 | 1/8w | 5\% | 4-7-470K | $\underline{\mathrm{H} 24}$ | 1 | 0.8 | 0.7 |
| C | 1/4W | 10\% | 4.7-10M | \% 12 | 1 | 0.8 | 0.7 |
| C | 1/2W | 5\% | 4.7-10M | E424 | 1-2 | 1 | 0.9 |
| C | 1W | 5\% | 4.7-10M | E12 | 2 -5 | 2 | 1.9 |
| MO | 1/2W | 2\% | 10-1M | H24 | 4 | 3.5 | 3 |
| WW | 1W | 10\% $\pm 1 / 20 \Omega$ | 0.22-3.9 | $\underline{\mathrm{EH}} 18$ | 7 | 7 | 6 |
| WW | 3W | 5\% | 12-10K | E12 | 7 | 7 | 6 |
| WW | 7W | 6\% | 12-10K | [12 | 9 | 0 | 8 |

Codes : $\mathbf{C}=$ carbon fllm high stability low noise
MO = metal oxide Electrosil TRS ultra low noise WW = wire wound Plessey
Valuen:
E12 denotes serien: $10,12,16,18,22,27,33,39,47,66,68,82$ and their decades E24: as E12 plua $11,13,16,20,24,30,86,43,61,62,72,91$ and their decades. Prioos are in pencs aqch for samo ohmic wodue and power rating, NOT mixed values. (Ignore fractions of ly on tofal value of resistor order).

## MULLARD polyester C280 series

$250 \mathrm{~V} 20 \% ~ 0.01: 0.022 ; 0.033,0.047$ 8p es. 0.068
 $1 \mu$ F 14 p . $1 \cdot 5 \mu \mathrm{~F}$ 81p, $2 \cdot 2 \mu \mathrm{~F}$ 84p.

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$0.33,0.39,9 p, 0.47,10 p, 0.56,13 \mathrm{p}, 0.68,15 \mathrm{p}$
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filled and pressurised. Terma: Dep. \&ll and 9 mathly. pyts. $88 \cdot 75^{(T o t a l} 871.75^{\prime}$ ).

R.S.C. AIO 30 WATT ULTRA LINEAR H|-F\| AMPLIFIER $\begin{aligned} & \text { Highly sensitive. Push-Pull high } \\ & \text { output, Hum level-70dB. Reaponse }\end{aligned}$ $\pm 318 B$
EF86, ECCB3, 807, 807, G234. Separate Baes and Treble Contrrols. EF86, ECC83, 807, 807, G234. Separate Baes and Treble Contrrols.
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For Mulard 510 Ammpliger $850-0-350 \mathrm{v}-100 \mathrm{~mA}, \mathrm{f} \cdot 3 \mathrm{v} .4 \mathrm{R}, 05-6 \cdot 3 \mathrm{v}$. 3 s $350-0-850 \mathrm{v} .160 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{~s}, 00-5-6.3 \mathrm{v}, 3 \mathrm{a}$.

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 $260-0-250 \mathrm{v}, 100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v}, 2 \mathrm{a} ., 6-3 \mathrm{v}, 1 \mathrm{~s}$.
$350-0.850 \mathrm{v}, ~$
$20 \mathrm{~mA}, 6 \cdot 3 \mathrm{v}, 2 \mathrm{~s}, 0-6-6 \cdot 3 \mathrm{v}$.


 $300-0-300 \mathrm{v} .130 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a} .$, c.t. $6 \cdot 3 \mathrm{v}$. 1 a . Guitable for Mullard 510 Amelifter $350-0-350 \mathrm{v}, 100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{~s} ., 0-5-6 \cdot 3 \mathrm{v} .9 \mathrm{am}$. | $350-0-350 \mathrm{v} .150 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{~m} .$, |  |
| :---: | :---: |
| $35-5-6-3 v . ~ 3 \mathrm{~m}$. | $28-20$ |

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wound EL34, 6L6, KT66 etc. to 3 or 15 8YOOTHIMG CHOKFs 150mA, $7-10 \mathrm{H}$. 250 . 80 $70 \mathrm{p}: 100 \mathrm{~mA}, 10 \mathrm{H}, 200 \mathrm{~S}, 60 \mathrm{p} ; 80 \mathrm{~mA}, 10 \mathrm{H}$, $350 \Omega 50 \mathrm{p} ; 60 \mathrm{~mA}, 10 \mathrm{H}, 400 \Omega \mathrm{Q} 26 \mathrm{p}$.
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(5) Patr 50W spenkers black
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$18^{\prime} \times 18^{\circ} \times 88^{-}$
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payments of 87.50 (Total $81 \mathrm{B5}$ ) payments of 87.50 (Total A185) FANE ULTRA HIGH POWER LOUDSPEAKERS
 High fux ceremie mas.s. continuons. 8
 $18^{\prime \prime} 100$ Watt $15^{\prime \prime} 60$ Watt 14.000 gauss $8 / 150$ 152.95

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FANE LOUDSPEAKERS 'POP' $25 / 212 \mathrm{in} .25$ WATT
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L12 12" 25 WATT 10,000 lines


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Plus master vol. control. Ind. Bass and Treble Plus master vol. control. Ind. Bass and Treble Controls. Protective circult to guard againat damage from accidental shorta. Output for Epeaker/s 3 to 30 ohms- 8ize $17^{*} \times 7^{\circ} \times 77^{*}$


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50
WATT AMPLIER A powerful high quallty all-purpose unit for lead, rhythm, bags guitar, vocalists. gram, radio, tape, Peak Output rating Joudspeaker unit optional horizontal or vertical mounting.太 Two extra hesvy dinty i2ln. Loudspeakers, * Four Jack inputs and two Volurae pick-ups or "mikes" Bass and Treble controls. Bend E.A.E. for lesilet.
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For Guitar, Vocal or
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A 4 input, 2 vol. control HI-FI unit with Geparate Bacs and Treble controls. Curreat vaives. Peak outpat raling. Btrong Rexlne black/gold P Y C. fecia Neon indictive For $200-250 \mathrm{v}$. A.C. malns. For 3 or 15 ohm Bend S.A.E. for leaflet. Terms: Dedoalt 54.50 and 9 monthly payments of 82.28 (Total ${ }^{1} 9 / 5$ Parr 2.5.02).



The Amtron range of products will shortly be available in the U.K. and a sole agent is required to handle distribution. These are iust four examples from their extensive range of equipment.


## Wide-Band Millivoltmeter

The measurement of small $A C$ voltages has assumed a great importance in the field of electronics. An essential requirement for instruments suitable for making these measuraments is that their insertion into the circuit shall not disturb the circult conditions. The UK430/A can be used for measuring background noise levels, residual ripple in power supplies and the measurement of frequency and gain in amplifiers.
Ratings and Characteristles: Voltage Ranges: $10-30-100$ and $300 \mathrm{mV}, 1-3-10-30-100$ and 300 V . Decibel Range: -40 to +50 dB . Frequency Range: 10 Hz to 3 MHz . Accuracy: $5 \%$. Input Resistance: 500 k ohms 10 mV to $3 \mathrm{~V}, 1 \mathrm{Megohm} 3 \mathrm{~V}$ to 300 V . Power Source: 9 V DC. Transistors: 4 x AF172, Diodes: $4 \times$ AA138.


## AM SIGNAL GENERATOR

This signal generator is a basic instrument for every technician, useful for AM radio receiver alignment, as well as for a wide range of measurements.
Radio Frequency Output voltage: 100 mV . Frequencies: 400 to 900 kHz , and from 950 to $\mathbf{1 6 0 0} \mathrm{kHz}$ Radio frequency attenuator: continuous variation. Modulation: internal at i kHz with depth of $30 \%$. possibility of exclusion.


## LOW FREQUENCY SIGNAL GENERATOR

Can be used to take various measurements, such as regulating equalizing circults of amplifler inputs, tuning filters to separate low frequencies from high in electro acoustic producing apparatus.
Frequencies: 10 Hz to 1 MHz . Max. output voltage; $1: 5 \mathrm{~V}$. Attenuator: 15 mV ; $150 \mathrm{mV} ; 1.5 \mathrm{~V}$. Output impedance: $\mathbf{2 0 0} \mathrm{Ohm}$.


## Level Meter

This instrument is very useful whenever there is a need to know the level of a slgnal belng fed into equipment. The unit is designed to be used in combination with the UK810 Dynamic Compressor which is provided with an output paint for connection to the ievel Meter The instru ment embodies a $200 \mu \mathrm{~A}$ full scale-deflection meter, a signal of 5 mV ampllude giving deflection.
Ratings and Characteristics: Input voltage: 5 mV max. Input Impedance: 47 k ohms. Power Source: 8V DC. Transistors: $2 \times$ BC108B, Diodes: OA95

Applications are invited from well established companies able to offer distribution facilities through out the U.K.
Please write with full details to:-
Box No. 102
Advertisement Dept, Practical Wireless, Fleetway House, Farringdon Street, London EC4A 4AD

## Up for the Cup!

$A^{\top}$T this time of year, all sorts of people are getting worked up on the outcome of that sporting event soon to take place in Wembley Stadium. In between bouts of waving their favours and whirling their rattles, the staff of P.W. have found themselves involved in the fate of another, though less spectacular, silver cup. They have been judging nominations for the P.W. Designer's Trophy for 1971.

In 1970 we launched Project Autumn, in which readers were invited to send in original articles on constructional projects within a specified period, the winner being the author of the work considered by the judges to be the best submitted. It was decided, however, to amend the rules slightly for the 1971 Trophy. Entries would be judged not in manuscript form but would be drawn from material actually published between the July 1971 and March 1972 issues.

Nominations would be judged, as before, on originality, ingenuity, construction, presentation, technical accuracy and other associated factors. However, under the revised arrangement, a new factor could be brought into playcall it 'workability', or Factor X. This allows the judges to take into consideration the trouble (or lack of it) in passing the articles through the editorial mill and noting any shortcomings not obvious in manuscript form. This permits a much fairer and realistic assessment to be made. The final results will be published in the next issue.

Although last year's crop of entries brought some new names to the pages of P.W., this year we want to see more and more! It has been suggested that the relative newcomer to technical writing has little chance when in competition with the regular and more experienced authors. There may, of course, be an element of truth in this, because on balance the "published" writer is experienced enough to have learned what is likely to be accepted for publication. On the other hand, P.W. has published many articles by new authors, sometimes at their first attempt.
However, in order to show faith in our stated objective of encouraging new writers, it has been decided that for the 1972 Cup only writers who have never previously had articles published in P.W. will be eligible. Describing the theory, construction and use of a piece of equipment can be a revealing exercise in testing the author's grasp of a subject (and the gaps in his knowledge), with the added incentive of some financial gain and, perhaps, a little glory! Full details next month.
So, keep those articles rolling-you may have just what we have been looking for and (who knows?) you may even win a cup!
W. N. STEVENS-Editor.

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[^2]
# NEWS... <br> NEWS... NEWS... 

## Pacemakers

With surgical care, an electronic module for a long-lived heart pacemaker is assembled at Raytheon Company's Industrial Components Operation at Quincy, Mass. Aerospace reliability techniques are employed in the manufacture of the new devices to pace the beat of a human heart. The new medical aids will be powered by nuclear energy giving them a life expectancy of


10 years instead of two as presently experienced with units powered by mercury batteries. Raytheon is building the modules for ARCO Nuclear of Apollo, Pa., a wholly-owned subsidiary of Atlantic Richfield. The experimental programme is sponsored by the Atomic Energy Commission.

About 50,000 persons now depend upon pacemakers. It is estimated that an additional 5,000 to 10,000 will join their ranks annually as the reliability and longevity of the device is increased.

Mr. Regan, Raytheon's marketing manager, said the electronic circuitry of contemporary pacemakers outlasts the present mercury power sources but with the shift to nuclear energy and its 10 -year life, the circuitry had to be upgraded. He said Raytheon had been selected for the task because of the company's experience and expertise in the manufacturing of sophisticated electronic modules. The company's performance with the ultra-reliability of the Apollo programme and compliance with NASA reliability standards were important factors in the selection, he said.

## 1,000W Amp

Crown International of Elkhart Ind. have announced a new high power amplifier capable of providing 1,000 watts r.m.s. into a 4 ohms load. Available from Macinnes Laboratories in the U.K. the M600, is d.c.-coupled throughout although a front panel switch provides for an a.c.-coupled input if required. Power bandwidth at the 1,000 watts level is $\pm 1 \mathrm{~dB}$ from d.c. to 20 kHz and THD at this level is below $0.1 \%$ throughout the frequency range. At 600 watts THD is below $0 \cdot 05 \%$. The intermodulation distortion is better than $0 \cdot 1 \%$ at all levels from $0 \cdot 01$ watt to 1,000 watts, and the damping factor is greater than 200 from d.c. to 20 kHz . Hum and noise are at least 110 dB below 600 watts level. The M600 is equipped with its own two speed fan for cooling

## "Radio Ear"

The present generation of mankind may be able to tune in on intelligent civilisations beyond Earth, if Congress appropriates the money to start building the VLA, America's proposed superpowerful radio observatory.

The National Science Foundation, under whose aegis the radiotelescope would be designed and constructed, said that an array of aerials resembling the giant "dishes" used to track spacecraft, would be set up in geometric patterns to pick up and focus signals emanating from the edge of the universe, 13,000 to 16,000 million light years away. The pattern would be an extension of the interferometer technique (placing aerials in a row or at right angles to one another) already widely used in radioastronomy to enhance the strength of incoming radio signals.

A specific location has not yet been chosen for the new telescope, but it would be built in a pattern some 24 miles long at a high-altitude site in the southwestern part of the United States well away from earth-generated background noise.
it is likely that any new very large aerial would be used, at least part time, in "passive listening" for radio signals from other civilisations which many scientists believe must inhabit planetary systems around distant stars.
and on the rear panel is a switch offering six positions of operating modes. Weight is 80lb., and price inclusive of import duty, is $£ 730$. Macinnes Laboratories Ltd., 71 Oakley Road, Chinnor, Oxon. Tel. Kingston Blount 52061.


The M600 amplifier

LM380 L.C.

D.T.V. Group Limited announce the LM380 audio power amplifier. The output has both short circuit current limiting and thermal overload current limiting to insure safe operation. Supply Voltage is 22V. Peak Current is 1.3 A . Power output is 2 W and input voltage is $\pm 0.5 \mathrm{~V}$. Storage Temperature is $-65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$ and Operating Temperature $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$. Lead Temperature (Soldering, 10 sec ) $300^{\circ} \mathrm{C}$. Price is $£ 1 \cdot 371_{2}$ +10 p postage and packing. D.T.V. Group Limited, 126 Hamilton Road, West Norwood, London, S.E.27.

## Gas Radio

Reader R. Jenkins has told us of a Mr. Jack Hickson who has a Gas-Powered radio.

The radio at Mr . Hickson's home in Hadnock Road, Monmouth, takes its power from a thermo-electric generator which has gas burners. Electricity is generated by heating of copper plates. The generator charges a .battery off which the 33 -year-old radio is run.

The generator supplies enough electricity to charge accumulator batteries or even a car battery.

The generator was made in 1938 by a Yorkshire firm specially for providing power for battery radios.

But for gas board engineers carrying out North Sea conversion it was something they had never encountered before. They had to take the generator to Cardiff and it took them about a week to change the burners and adapt it for North Sea gas.

## Audio Fair

Exhibitors at last year's International Audio Festival \& Fair asked for more space in the 1972 show. In order to accommodate them, the organizers have arranged to move the Festival from the Empire Hall to the Grand Hall, Olympia, which will provide an additional 74,000 sq. ft. of space.
The 1972 Audio Fair, sponsored by the Sunday Mirror, will be held from October 23 to 28. It will allow the public to listen to the products of the world's leading manufacturers of sound reproduction equipment, specialist Hi Fi , and a comprehensive display of accessories.
There will be many special attractions, including more than 75 specially constructed Audio Studios, and a Hi-Fi Theatre offering a daily programme of lectures and concerts.

## Gais Light



GaAs Light indicates the state of p.n.p. or n.p.n. logic systems capable of driving at say 15 mA at 1.6 V or over. For the brightest display 40 mA is recommended. The intensity is acceptable at only 10 mA . The advantage of these lamps is the low cost (95p) and higher reliability.

Due to the low current and voltage requirement of small battery operated equipment it is now possible to run these GaAs lamps in series with the electronics, so reducing the indicating power to only 16 mW . This power is therefore less than the power wasted by even a very small equipment being inadvertently left on. West Hyde Developments Limited, Ryefield Crescent, Northwood Hills, Northwood, Middx., HA6 INN.

## Led Lamps



Guest International announce the availability of a new range of solid state lamps using Gallium Arsenide Phosphide Diodes. The lamps are encapsulated in two main forms of package-one with a glass to metal seal and the other a moulded plastic case. Both employ a lens in front of the actual diode to alter the characteristics of the light emission, one type diffuses the light, whilst the other produces a point source each available in different colourations. The lamps combine extremely long life with very efficient operation making them ideal for use as panel lights or status indicators in modern equipment. There are currently 5 versions in the range given the generic title RED LIT series and luminosities as high as 2000 FtL at 50 mA are offered. Guest International Limited, Nicholas House, Brigstock Road, Thornton Heath, Surrey, CR4 7JA.

## Philips Booklet

Philips have published a colour booklet entitled "All about Tape and Cassette recording". The booklet, which sets out the origins of tape recording and describes how magnetic tape works, is available to schools and students on request.

It has been published primarily as further study material for project work.

25,000 copies of the booklet have been printed and copies can be obtained from the Publicity Department, Philips Electrical Ltd., ELA Division, Century House, Shaftesbury Avenue, WC2H $8 A S$.


SINCE the orginal Treasure Tracer was built and described in this magazine, development work has continued. The author's own success with the prototype and the enthusiasm of readers has resulted in a considerable amount of experimenting and this article describes the improvements that have been made.

It has not been assumed that all readers will have seen the original article; for this reason much of the information given last time is repeated. This article should enable the. new reader to build the project as described as well as enabling the reader who built up the original circuit to up-date his Treasure Tracer. To make matters simpler for these people, the original case has been retained with relatively few changes.
The cost of the original was not high but in fact the component cost for the Mark II will probably be marginally less, though the mechanical and construction side may cost more. The total cost for the Treasure Tracer Mk II should be something under $£ 5$ though this will vary depending on source of supply.
For those who did not see the original article, the principle of operation will be briefly described.
The circuit works in exactly the same way as the original. Two low frequency r.f. oscillators are arranged to beat together. One of these oscillators makes use of a tuned circuit working inside the case; we shall call this the reference oscillator. In operation this remains at one frequency though this is adjustable for "tuning up".
The second oscillator, working at essentially the same frequency, also makes use of a tuned circuit but in this case the inductor is external and becomes the "business end" of the circuit. We shall call this the search coil and the search oscillator.
If a small part of each of these oscillators is mixed together they will beat with each other. If the difference in frequency is small, this beat note will fall in the audio range. If the reference oscillator is working at say 130 kHz and the search oscillator is at 131 kHz , a beat note equivalent to that difference will be produced in this case 1 kHz ; this audio signal can be amplified and fed to the speaker.
Now the search oscillator makes use of an external coil and if certain materials, especially metals, are brought into the vicinity of this coil, these will alter

## The original Treasure Tracer, described In P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This articie describes an improved version of this <br> The original Treasure Tracer, described In P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This articie describes an improved version of this <br> The original Treasure Tracer, described In P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This articie describes an improved version of this <br> The original Treasure Tracer, described In P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This articie describes an improved version of this <br> The original Treasure Tracer, described In P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This articie describes an improved version of this <br> The original Treasure Tracer, described In P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This articie describes an improved version of this <br> The original Treasure Tracer, described In P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This articie describes an improved version of this project.

## HALVOR MOORSHEAD

the value of the coil inductance, which in turn will alter the frequency of oscillation.
A better known example of this change of inductance is the ferrite rod aerial. By moving the coil along the rod the inductance is changed; this is of course done every time a superhet is aligned. A ferrite rod will roughly double the inductance of a coil. Far smaller changes in inductance take place
in a metal locator-usually fractions of one percent -but because of the frequency used and the fact that it is compared to a standard frequency, the changes in the audio note will be considerable. Taking the frequency actually used in this circuit, a 0.1 per cent increase (or decrease) in the inductance will result in a 130 Hz change in the audio note.

It would be possible to have both frequencies fixed at say a difference of 100 Hz but there are so many factors that effect the frequency that it is easier to arrange for one of the oscillators to be adjustable in order that they can be brought close together.
In this circuit we are changing the frequency of the reference oscillator though it could just as well have been the search oscillator.


## MODIFICATIONS

In the light of experience and bearing in mind the comments of readers, the following modifications have been made to the prototype:

1. The original Treasure Tracer was not mechanically balanced. It was held at the end of the handle and while this worked quite well, it became very tiring to hold during long searches. The Mark II is balanced as can be seen from the photograph; it is easier and more comfortable to hold and there is no hand strain even after several hours.
2. The output volume was felt by some readers to be too low. In many cases this could have been due
to inefficient operation of the particular transistors used as the output of the prototype was found to be quite adequate. Even so, the sound level available in the Mark II has been increased and since this could be annoying in certain locations at maximum volume, a volume control has been included. To save extra drilling etc., the on-off switch (a miniature slide type in the prototype) is incorporated with the volume control.
3. The stability of the prototype left a little to be desired. Touching the metal case or the wires running from the search coil to the case caused a change in frequency. This was not too serious in practice once one got used to it but several changes have been made to the construction of the Mark II to improve the stability.
4. The construction of the search coil has been changed to make the windings much more rigid; this has proved to be much better than the principle originally employed. In addition the weight of the "head" has been reduced and this again helps with the balance.
5. Overseas readers (who account for a proportion of the magazine's readership) sometimes had diffculty in obtaining certain parts. This has been borne in mind and the parts should be available world-wide except for one component and the construction of this is described later.
6. Improvements have been made to the oscillator circuits. In the Mark I a Hartley osciliator and a Blocking oscillator were used for the search and reference oscillators respectively. Both of these have been changed to the more stable (and slightly simpler) Colpitt's type. In addition to these advantages, the search coil is now only a single winding (it was tapped in the Mark I) as is the coil used for the reference oscillator. This simplifies construction and reduces the cost slightly.
In addition to the above points a considerable number of other experiments have been tried but have not been incorporated for various reasons. For those who wish to experiment further, these are mentioned; they may give certain leads and possibly prevent the reader following fruitless paths.
The best frequency for operation was carefully investigated. Although the regulations limit operation below 150 kHz , experiments were carried out at six widely differing frequencies and the results compared. Since for a given percentage change in inductance there will be a much larger change in the audio frequency using higher r.f. frequencies, it would appear that operating at say 2 MHz would be better. In practice other factors start to come into it and ground capacity effects become serious. Bringing the coil towards plain earth causes a considerable change in frequency and even fitting a Faraday Shield only improved this slightly. About six different frequencies were tried but the only result was a confirmation that the original 130 kHz or thereabouts is the best. Ground capacity effects are negligible and, although a Faraday shield was tried, this made no improvement and so has not been incorporated.
A meter was tried as the sensing indicator. Although a workable circuit was developed, it was found to be far less sensitive than the ears and very difficult to use. There was also considerable difficulty in filtering out all the r.f. before applying this to the frequency counting circuit. For this reason details are not included.
Several experiments were made with different diameters of search coil. A clear rule seems to apply
here. The larger the diameter of the coil, the greater the depth of penetration for large objects but at the same time the larger coils were less sensitive to small objects close by.

There is certainly room for experiment here. As long as the inductance of the coil remains the same, it doesn't matter at all what size it is, it could be lin. or 3 ft . in diameter. The 6 in . used here is the best compromise for general purposes; it is good for coins and not too bad for large objects reasonably close by For those who are prepared to take the trouble, it should be possible to arrange for the coils to be changed depending on the nature of the search.

As a general guide the same length of wire should be used whatever the diameter of the coil, that is about 92 ft . of wire. This is only a guide and turns may have to be added or taken off, to allow the frequency to be brought into the right range.

Encapsulating the coil in epoxy resin was tried and although this was an improvement over the Mark I as it prevented movement and consequent frequency changes, it did not help with the Mark II and only added to the weight of the search coil head.

## THE CIRCUIT

The complete circuit is shown in Fig. I. Trl and the components surrounding it form the search coil oscillator and, as can be seen, this connects to the search coil via a coaxial cable. $\operatorname{Tr} 2$ and the associated components form the reference oscillator, L2 being a long-wave winding (an inductance of just over 2 mH ) but fitted with a ferrite slug (not normally supplied). The one used in the prototype is primarily intended as a crystal set coil; a medium wave winding is also fitted but this is ignored.
This oscillator is adjustable by varying VC1 which is a 750 pF compression trimmer and the combina-


Fig. 1: The complete circuit of the Mark II. Although very similar to the original as far as operation is concerned, this is more stable and has a greater output.

## components list

## Resistors

| R1 $470 \mathrm{k} \Omega$ | R5 | $120 \mathrm{k} \Omega$ |
| :---: | :---: | :---: |
| R2 $10 \mathrm{k} \Omega$ | R6 | See text |
| R3 $470 \mathrm{k} \Omega$ | $R 7$ | $270 \mathrm{k} \Omega$ |
| R4 | $10 \mathrm{k} \Omega$ | R8 |
| R | $39 \Omega$ |  |

All tesistors $1 \mathrm{~W}, 5 \%$ types.
VR: 10 k ) log pot with switch

## Capacitorg

|  | 2000 : cepanic of polystyrene |
| :---: | :---: |
| 2 | 2,00ar centmic or polystyrene |
| C3 | 0.054 F certinic or my |
| C4 | 3,300, F caramic or polystyrene |
| C5 | 005 F eersmic or mylar |
| cs |  |
| c7 | 2twne |
| C8 | 3, 300 pf ceramic or polystyrene |
| C8 | Q ${ }^{4} 4$ F ceramic or mylar |
| C10 | atpe 10V electrolytic |
| C11 | $50 \mu \mathrm{~F}$ 10V electrolytic |

## Semicondiuturs

| Tr1 | 248080 | Tr4 | BC109 or 2 N 2926 |
| :---: | :---: | :---: | :---: |
| Tr2. |  | D1 | +N914 |
| Tr3 2t 20.26 |  |  |  |
|  |  | be | e 2N2926 can be for Tr3 and Tr4) |

## Miscollaneous

W. Side Eext and Fig. 2

L2. Reponco Crystal Set Coil Type DRX1. This sioust be fitted with a ferrox tuning slug which I 1 not provided.
4. tumsistor output transformer, Type LT700 or slmintr-see text.
4S. Ningature 80 loudspeaker-see text
Earphople jack, 3.5 mm type with break switch; सittery; $\mathrm{PPB}_{3} 9 \mathrm{~V}$; Case, $6 \frac{2}{2} \times 2 \times 1 \mathrm{in}$. in aluminium quallable from H. L. Smith Ltd., 287/9 Edgware Rad, London W.2, price 60p including postage); Coax cable; Connecting block; Aluminium for croper) tube, 54t; Plywood for former.
tion of the variable core in the coil and VCl enables this tuned circuit to be varied over a considerable range. Even if the search coil oscillator is way off the intended frequency it will be possible to tune the reference oscillator to it.

From the collectors of both oscillator transistors low value capacitors C6 and C7 feed to the detector diode which, being a silicon type, requires a standing
d.c. bais provided by R5. The output from this is fed to the volume control VR1 and is smoothed by C8 which removes nearly all the r.f. component. The output of the volume control feeds to the simple amplifier which uses $\operatorname{Tr} 3$ and $\operatorname{Tr} 4$. R6 has not been given a value as it should be selected individually; its value will depend on the actual transistors used. For initial setting up a $22 \mathrm{k} \Omega$ resistor may be used,
this will certainly work. Once operation of the main stages has been established, the correct value should be found by experiment. It should be chosen for maximum output compatible with reasonable current consumption (under 20 mA ).
Some readers had difficulty in obtaining the high impedance speakers used in the Mark I and for this reason the Mark II uses a standard $8 \Omega$ miniature loudspeaker. This necessitates using a transistor output transformer. Normally only push-pull types are available, that means that the primary is centre tapped, in this case only one half of the primary winding is used. The use of earphones is also easier as the impedance of $8 \Omega$ is far more common than the $80 \Omega$ used in the Mark I. For those who have high impedance speakers (anything from $35 \Omega$ to $80 \Omega$ ) these can be used, in which case they go directly between the collector of Tr 4 and the positive line. The fitting of an earphone socket for this will have to be worked out if required. R6 will be a different value if a high impedance speaker is used.
The whole circuit is operated by a PP3, 9V battery which will last a considerable while. The current consumption will depend on the value of R6 but will probably lie in the range $10-20 \mathrm{~mA}$.
For those who have the original Treasure Tracer article, it will be seen that although very similar in many aspects, this circuit is rather simpler as well as being a bit better. The choice of a BC109 transistor for $\operatorname{Tr} 4$ rather than the 2N2926 used previously is not important and a 2 N 2926 can be used here if required.

## CONSTRUCTION

The construction of the Treasure Tracer Mark II can be regarded in two distinct sections; the electronics and the mechanics. It does not matter which is tackled first.
The search coil is wound on a wooden former cut from a piece of ${ }^{1} 4 \mathrm{in}$. ( 6 mm ) plywood, see Fig. 2. This is circular in shape with a cut-out to reduce the weight although this is optional; it is a simple matter with a fret-saw or jig-saw but difficult otherwise. A groove should be cut into the edge of this circle to hold the wires. Many methods can be used to do this, the simplest way is to spin it in an electric drill and use a rasp to cut the groove. The search coil need not even be round, it could be square or triangular and many other methods of construction can be employed but the described shape has proved very successful. When this groove has been cut, two very small holes should be drilled diagonally from the groove to the top of the former as shown. Where these emerge, a connecting block can be screwed on to act as the winding anchoring point. Sixty turns in all are required; these should be neat and taut and once finished a band of tape can be wound around as shown in the photograph. An alternative would be to saturate the windings with a glue to hold them firmly in place. By sinking the windings in a groove they are protected to a considerable extent.

The handle of the device is made from aluminium tubing which is available from a number of sources. If difficulty is experienced in obtaining this it can be made from copper tubing with the same diameter, a plumber should be able to supply a suitable length. Five feet was found to be about the right length but this will depend on the height of the user.

The handle is fitted to the search coil former with a simple bracket, bolt and wing-nut as shown in Fig. 2. The connecting wire from the search coil to


Fig. 2: The construction of the search coil. This design has been found to be excellent but as long as the inductance of the coil remains the. same, any arrangement may be used.


A photograph of the search coil used in the Mark II.
the main circuit is a coax cable running up the tube. At the lower end this is fitted to the connecting block. The use of coax ensures that there will not be loose wires flopping about inside and that the capacity will remain constant. Coax is more rigid than twin wires and is also better for that reason.
The electronics is built into a aluminium case, available from the suppliers mentioned in the components list. This will have to be drilled to take the speaker and various other fittings. A drilling diagram is not shown as the siting will depend on the components used and in any case the siting is fairly obvious. A series of small holes can be drilled for the speaker or a large hole ( $1^{1}{ }_{2}$ in. diameter at least) which can be fitted with some form of grille.
The majority of the components are mounted on a piece of Veroboard and the layout of this is shown in Fig. 3. Veropins are used for all the take-off points as this leads to a nice, neat appearance. The coil L2 is mounted directly onto the board as shown. Four corner mounting holes should be drilled for later fitting.

A number of breaks are needed in the copper strip of the Veroboard and these are indicated in Fig. 3. There is nothing complex about the wiring. R6, which will have to be found by experiment, has plenty of room around it. The easiest way of experimenting with this is to initially wire in the recommended $22 \mathrm{k} \Omega$ resistor with fairly long leads. Once the general operation has been established this can be cut out leaving wires sticking out. These can form the anchoring point for the replacement.
Figure 4 shows the remainder of the wiring and also the anchoring of the coax cable from the search coil. All connecting points are coded with letters and no difficulty should be experienced in this part of the construction.

(x) Mounting holes


Fig. 3: The components layout on Veroboard. Veropins should be

## TESTING

Once completed the unit can be tested. If all is well there will be a slight noise from the speaker which will be increased greatly by dabbing a finger onto the VR1 contact which runs to point A on the board. Set VCl to roughly half compression and adjust the core of L2. A strong whistle should be heard at one point and this will be found to be tuneable by VCl as well. Other lower level whistles may


An internal view of the Treasure Tracer Mark II.


Fig. 4: The wiring inside the case. VC1 must be sited so that it does not foul the coil $L 2$.
be heard, these are due to the oscillators beating with radio stations. In some cases it is possible to make the unit operate using only these but it will not be nearly as loud or as good. The level of the main beat note will be much louder and cannot be missed.

Once the beat note has been established, the Treasure Tracer is ready for use. We have deliberately avoided making any claims about the depth to which it will find objects. Commercial models frequently make wild claims which cannot be substantiated and we do not want to follow this example.

## To conform with the Wireless Telegraphy Act (1949) a licence is required to use the Treasure Tracer described here.

Under Section 1(1) (Pipe Finder Licence) the band 16 to 150 kHz can be used for equipment of this type.
A licence for five years costs 75p and can be applied for on a form obtainable from the Ministry of Posts and Telecommunications, Waterloo Bridge House, Waterloo Road, London S.E.1.


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## DEPTH OF OPERATION

In the original article we showed exactly what we found and at what depths. Generally speaking coins can be detected at ranges of up to 4 or 5 in . but this also depends on the size. Large objects can be detected at distances up to about a foot but the sensitivity falls off considerably over that distance.
Do no expect to become an expert as soon as you start to use the Treasure Tracer. The Mark II is considerably more stable than the Mark I and there will be fewer false signals; even so it takes an appreciable amount of practice before the small changes in the audio frequency are noticed. This point cannot be over-emphasised.
Using the Mark II one person was able to locate a ${ }^{1}{ }_{2} p$ coin at 4 in. every time, while another, who had never tried it before, had difficulty in finding a 50 p coin at 2in. This illustrates the importance of practice.
The beat note is adjustable for either a rise or fall in frequency when a metal object is approached. The author prefers a falling note but this is only personal preference. It is usually better to work with a low frequency audio note-say 100 Hz -as a 50 Hz change at this frequency is very noticeable; at 1 kHz the same change would not be so obvious.

## FREQUENCY LOCKING

If the two notes are very close together, the two oscillators will lock together. The Mark II is slightly better than the Mark I in this respect and reducing the value of C 6 and C 7 will lower this locking frequency even further but no advantage was found in bringing it much lower. The prototype Mark II locked at about 20 Hz .

VC1 has considerable range and it has been found better to arrange for it to be used at near maximum compression, compensating for this by readjusting the core of L2. This prevents a change in frequency due to the leaves of VCl moving which they tend to do when there is little compression. A possible improvement would be to replace VC1 with a much lower value-say around 100 pF (compression trimmers in this range are available) and to wire fixed capacitors in parallel with it to obtain the correct value. This will mean that VC1 will have a narrower range, but still around the correct frequency, at the same time as reducing the likelihood of instability.

L2, as we have said, is a long-wave winding of a crystal set coil, the only modification being to fit a ferrite core (the former is already tapped to take this). It is possible, if tedious, to wind one's own coil on a ${ }^{1}{ }_{4}$ in. outside diameter former; 350 turns of fine wire ( $39 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. is used by the makers) pile wound will give the correct inductance.

One point must be repeated from the previous article. Users of metal locators are not very popular with many archaeologists because of the highly irresponsible way in which some people use theirs. A local archaeological society may be happy to use your services under their supervision but others may not welcome you. Note that there are heavy penalties for using such a device in areas scheduled as being of historical interest without permission.

[^3]MANY people have expressed a mixture of interest and scepticism about the 50F (fifty Farad) "capacitors" which are contained within $1_{3}$ rd of a cubic inch. These are manufactured by Gould Ionics in the U.S. and are marketed in the U.K. by Lyons Instruments at Hoddesdon. Note: the price of these units for one-off is around $£ 25$ each and the company may well only supply the professional user.

How does your radio shop test those transistor radios and televisions? Bet they have an engineer prodding about with a multimeter. Of course, knowing exactly where to prod and why is the most important part.

In industry, it would take far too long to do this on the end of a fast-moving production line so other techniques have had to be employed. Automated test gear is a highly sophisticated subject, but it is becoming more commonly employed.

One piece of gear which does this has been introduced by Honeywell Limited. Designated the "Swift" digital tester, it is a portable equipment and is so simple to operate that even an unskilled person can use it successfully. It can be used for testing digital printed circuit boards when they are fully assembled. Amazing thing is the speed at which the tests can be carried out. The equipment can test at the rate of a million tests every second which is just a trifle faster than the average service engineer. A p.c.b. with some 64 connectors may be plugged in and tested and should a fault arise, this is indicated by a light which shows at which end connector the fault is.

Cable-Scan is another useful little test gadget marketed by Thomas \& Betts International Inc. located at Cowcross Street, London. Imagine you had a bundle of wires to sort out. Which are the signal wires, which are earth etc? Well, Cable-Scan will tell you.

A clever idea is that the operator wears a small wrist strap which provides a signal. This is then cunningly transmitted via the operator's fingers so that all he has to do is to touch the wires. A digital readout is provided so there is no chance of an operator making an error through having to write data down-he can see at a glance. Up to 200 signal input circuits are catered for which means that the operator can tackle a mighty big bunch of wires and sort them out without any trouble.

Before you think of rushing out and buying an automated tester and setting up in business as a lightning radio and t.v. repair enterprise, the cost of the Honeywell Swift equipment is around $£ 900$. But it is still cheap for large production runs.

## The "colonials"

I am writing in hope that you may like to enlighten "us Colonials" about some of your circuits that you print in that great little magazine P.W.

We in Australia cannot obtain quite a few components that you use, mainly coils and substitute transistors. Wouldn't it be possible to give information on coils, such as diameter, turns, gauge of wire, etc.
There is another favour you can do for your readers, don't change P.W., it's O.K. as it is.R. C. Shambler, (Australia). Editor's Note:

It would not really be practical to give all the information on the coils specified in P.W. However the majority of the larger advertisers in this magazine will supply components to overseas readers and even air mail postage on small items such as coils and transistors is surprisingly inexpensive. (To Australia the air mail postage is often only twice that of internal British mail).

## Cassette deck

I agree with V. C. Watts (Bath) "Cassette deck" March '72. I have tried many companies in search of a deck for a home-made system. The only firm I have discovered who at one time did sell decks (made by Philips) has now stopped and according to their letter did not think they would be able to sell them again. Perhaps these and other letters may stir the component firms into action.-A. R. Knight, (Oxford).

## Ebuc oidar

Leafing through my March copy of Practical Wireless I chanced upon the article on the Cube Radio. On closer inspection, I found that the photograph on page 998, a general view of the completed receiver, was shown as a mirror image. The lettering on the battery and the position of VR1 show this. Is this because we have to use a battery providing positrons-the mirrorimages of electrons?-G. L. Manning, (Middlesex).

Alice is not a member of our Staff-it's all.done by mirrors(Editor).

## Tools for the job

The purpose of this letter is to comment on "Who makes What?" In Feb. issues's "Letters", as it has some bearing on my own experience, although with ancillary tools rather than radio parts.

Last May I got a kit for the "PW-12-12 Stereo", which is now, after some odd troubles working fantastically well. I was determined to get the tools for the job, which included special soldering irons and bits, the standard 'nipper' type of wire cutting pliers, and long nosed ordinary pliers, I was also determined to get British tools which I did; or thought I did. My comments hereunder refer to the pliers and wire nippers.

They were both on the modern type of display card, the long nosed pliers, with nicely red painted handles I saw were made in England. Quite a number of filing marks here and there on them, and a scratchless finish which looked like chrome plating. These nippers had the name in large headlines of the English company which 'made' them as I thought. A London firm. Surprisingly despite the perfect finish and insulated grip of the 'nippers' there was only $10 \%$ difference in the prices, and I thought that British tool makers at last were 'With it!'

When I got home I found that these perfect nippers were made in Japan, as quite properly announced in tiny print, for the English firm, and I felt aggrieved.

So I went to work on my Double-12 amplifier, and exactly a month later I found that the English made pliers had ${ }_{166}$ in lateral play at the plier points, showing either bad workmanship or bad material. After nine months the Japanese nippers are as when bought and when I'm sewing on a trouser button, I use them in preference to scissors to cut the cotton! Clean as a whistle.

I know this does not refer to radio parts, but such an experience reflects on all British products, and I wonder if this is the reason why it is now apparently expedient not to put the country of origin on the product. -N. H. Hodgson, (Middlesex).

## CQ thanks

I should like to thank you for printing my problem in your C.Q. column. I have had several replies recently, including one offer from a Mr. Oldknow in Horsley, Woodhouse who is willing to turn a new pulley wheel from scratch on his lathe!-W. R. Kennon, (Derbyshire).

## Dreams

I often dream of being a ship's radio operator between the sparkgap and the valve era, with big brass morse keys, large accumulator terminals, 4 pin valves and coils on 2 in formers and not ${ }_{4}{ }_{4}$ in. The transistor should never have been invented. Old and sentimental at 26 years of age!
Just in case I start a war as. an inexperienced nut perhaps the following details provide my defence. My construction knowledge covers from a crystal receiver to a 9 valve communications receiver.
Operating experience $-42,000$ miles of travel.
Operating a.m., s.s.b., c.w. and RTTY links plus Australia with the armed forces, presently employed as GPO Telegraphist.

Transistors are too small and don't leave enough room for any serious experiments, I'll still keep buying old battery valve sets for 50 p and $£ 1$ and re-making them until they run out of supply and then I'll take up fishing.-W. D. Logan, (Hyde, Cheshire).

## Nuvistors

In "Letters" of March, Mr. K. Freeby remarks on the absence of information on nuvistors. I have a copy of Practical Television dated December 1962 in which there is an illustrated article by K. Royal entitled "Practical Nuvistor Circuits". Mr. Freeby may have this magazine if he requires it.-N. A. Hodgson, ( 84 Lansbury Drive, Hayes, Middlesex).

We would like to thank $M r$. Hodgson for his kind offer and also to thank Mr. Laurence Mason who has kindly sent us some gen on nuvistor valves to pass on to Mr. Freeby-Editor.

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YOU want to listen to Radio 2 but you haven't got long waves or v.h.f. on your transistor portable. What to do? Just place the Add-aVerter behind your radio and tune in Radio 2 at about 480 metres on the dial of your receiver. No connections or fiddling about with the innards of your squawk box.

How is it done? By picking up the Radio 2 signal on 200 kHz ( 1500 metres) and mixing it with a local oscillator 429 kHz in the Add-a-Verter thus producing a new signal on 629 kHz which can be heard on the receiver on the medium wave band.

Cost of the bits and pieces bought new? About $£ 2$, plus the cost of the plastic box. The average constructor will be able to find most of the parts in his junk box.

## CIRCUITRY

Nothing very complicated as can be seen in Fig. 1. A commonly available 2N2926 n.p.n. silicon transistor has a quartz crystal connected between its base and collector with the supply voltage being applied via


Fig. 1: The simple circuit of the Add-a-Verter. The output coil $L 2$ is coupled to the ferrite rod aerial in the receiver, as shown in the heading photograph.

## 

L3 which presents a high impedance at the crystal frequency. The circuit will still oscillate if the bottom end of C1 is connected to the negative line. But more later on the operation of the crystal.

The Radio 2 signal is picked up on the ferrite rod aerial L1 and fed to the base of the transistor. The signal at the collector now consists of the oscillator frequency 429 kHz plus and minus 200 kHz as well as the basic oscillator frequency of 429 kHz . However it is the sum frequency of 629 kHz which interests us, since it falls in the m.w. band.

The output coil L2 is also in the form of a ferrite rod aerial and it is the signal from this coil on 629 kHz that is picked up by the ferrite rod aerial in the adjacent receiver.

## THE CRYSTAL

The use of a crystal may seem extravagant and unnecessary to some but in fact it costs just 25p and its use saves much time fiddling around with coils and capacitors trying to hit the right frequency. Its stability is also so very much better than an LC combination. It is worth noting that the crystal oscillated immediately the prototype was switched on for the first time. The only other adjustment then needed


The only 'control' on the Add-a-Verter is the on-off sw/tch!
was to peak trimmer TCl for maximum strength of the Radio 2 signal.

This particular type of crystal was originally used in equipment that multiplied its basic frequency fiftyfour times so the crystal is designated as having an output frequency of 23.2 MHz which is shown on the crystal case as Channel 32. In fact its fundamental frequency is 429.630 kHz . There is a whole range of similar crystals between about 370 and 540 kHz . Some extremely useful tables showing the relationship between channel number, fundamental and output frequency can be found in the Radio Data Reference Book published by the RSGB.
To avoid interference with the converted Radio 2 signal from other stations on the medium wave band a search was made for a "quiet" spot at the l.f. end of the band but such a rarity disappeared a long time ago! The choice of the final frequency is therefore something of a compromise. What may seem a relatively quiet spot on the band in one part of the country may be alive with stations in another. However the Radio 2 signal appearing at the receiver was strong enough (in Surrey) to eliminate any other stations that can normally be heard at night on 629 kHz . During the daytime there is no problem at all.

## CONSTRUCTION

First obtain a plastic box, with lid, about $4 \times 3$ in. with a minimum depth inside of about $1^{1}{ }_{4} \mathrm{in}$. The inside dimensions of the box will determine the size of the board on which the Add-a-Verter is built. The box used in the prototype was $4^{1} \times 3 \times 1^{1} 4 \mathrm{in}$. so the following description and accompanying diagrams are based on this.
The important factors to be observed, regardless of


Fig. 2: Layout of the few com-


Board $4^{\prime \prime} \times 2 \frac{3 / 4}{4}$ but see text
(CKG107)


Top view of circuit board, which may be compared with Fig. 3
the size of the board, are maximum separation between L1 and L2 while keeping them parallel and horizontal and keeping L3 away from both L1 and L2 to minimise interaction.

Cut a piece of plain Veroboard to the size required and insert Veroboard pins as shown in Fig. 2. The use of copper strip board was considered unnecessary and a waste with so few components to be mounted. Coils L1 and L2 are part of a m.w./l.w. ferrite rod aerial although any similar rod aerial can be used.

First remove the coils from the rod and then cut off two pieces of the rod ${ }^{2}{ }_{4} \mathrm{in}$. long. If space permits simply cut the rod in half. The cutting is best done by nicking the rod at the appropriate place with the sharp edge of a file. Either hold the rod in a vice and tap the free end or strike the rod smartly on the sharp edge of a table or bench. If the resulting ends are a bit rough, don't worry, it won't make any difference!

The connections to the coils L1 and L2 are shown in Fig. 4. L2 should have the few turns constituting the base winding removed although this is not strictly necessary provided the ends of this winding are not allowed to short together. If difficulty is experienced in identifying the windings connect an ohmmeter to


Fig. 4: Diagram of the lead-out wires of coils L1 and L2.
the single lead "a" and find the other end of the winding " b ". The remaining ends will be the base winding.

Put the coils on the ferrite rods and attach the rods to the board with loops of thin insulated wire. Drill a hole and fix the compression trimmer TC1 to the board with the lock nut provided. Place the tags of the crystal holder over the appropriate pins on the underside of the board and solder, see Fig. 2. These pins provide anchoring points for the base and collector leads of the transistor on top of the board.

Coil L3 is the long wave winding of a l.w./m.w.
coil, freely available, and intended for use in a simple "crystal set" circuit. An iron core to suit the coil is required and should be bought at the same time as the coil since the core is not normally sold as part of the coil. In the prototype the red and blue tags of the l.w. winding were soldered directly to the appropriate pins in the board, Fig. 3, but if height permits mount the coil normally by its fixing lugs. Initially the core should be screwed in as far as possible.
Rather than cut off the excess of the coil lead-outs it is advisable to use the tinned ends as supplied, winding any surplus wire back on to the coil formers.
Complete the wiring in of the components as shown in Figs. 2 and 3 leaving the transistor until last. Hold the leads of the transistor in a pair of flat nosed pliers while tinning the ends and soldering them to their pins. Transistors are pretty tough these days but there is no point in tempting Fate!
Finally, attach the PP3 battery to the board with a wire loop as was done with the ferrite rods. The negative side of the battery clip goes to the general negative line while the positive side goes to the on-off switch, mounted on the lid of the box, and back to the positive side of the circuit. The current taken by the prototype was only $120 \mu \mathrm{~A}$ so there was a temptation to omit the switch altogether! In this case the


Photograph of the underneath of the circuil board with conneclions as shown in Fig. 2.
working life of the battery would be about the same as its shelf life.
Plug in the crystal which should lie flat against the board but if necessary fix it to the board with another loop of wire.

## SETTING UP

Put the board into the case where it should lie flat on the case of the crystal. No additional fixing was thought to be necessary. Connect a milliammeter in series with the positive battery lead. If using a multimeter set it to the maximum current range in case there are any short circuits! Switch on and if the crystal oscillates first time the current will be of the order of $120 \mu \mathrm{~A}$. If not the current will be about $200 \mu \mathrm{~A}$ when the core of coil L3 should be unscrewed slowly the current dropping to a minimum and rising again. As a final check on the crystal, switch the unit on and off a few times. Each time the current should fall to around $120 \mu \mathrm{~A}$ but if not readjust the core until it does. To be quite sure, remove the crystal, when the current should rise.

Switch on the portable radio and set the dial to

## components list

> R1 4 NM 62 fw .
> C1. 1000
> C2. 100 pF
> C3 t 100 p e fif raquired)
> C4 0tara 12 VW
> TC1 250 DF compression trimmer
> Tí 2 N292s (orange, yellow or green coding)
> L1/L part of ferlfer rod aerial (Type MW/LW/FFR Denco)
> 43 part of LW fm.w; coil, with core (Type DRX1 Henrys)
> X1 crystal type FT 241 A , approx: 429 kHz ("Channel $32-23-24 \mathrm{C}^{\prime}$ ", Henrys) and folder
> Miniature onfoll switch; PP3 9V battery and clips.
> *Veroboard 0 -15in matrix or plain paxolin, $4 \times 2 \mathrm{in}$ (see text) Plastic box.
around 480 metres on the medium wave band. Stand the Add-a-Verter close behind the radio so that the output coil is close to the ferrite rod aerial in the set and parallel to it. There is a fair chance that Radio 2 will now be audible but if not tune around 480 m . on the set at the same time adjusting trimmer TC1 in the Add-a-Verter until Radio 2 is heard. Adjust the trimmer for maximum volume. To check, move the Add-a-Verter away from the set when the Radio 2 signal should disappear. For portability the Add-aVerter can be attached to the radio with a couple of rubber bands.
Should the m.w. receiver not have ferrite rod aerial the Add-a-Verter can still be used by adding capacitor C 3 from Trl to a terminal on the box and joining the terminal to the aerial socket on the receiver.

## NOTES

Do not use a metal box since operation of the Add-a-Verter depends upon the coupling between coil L2 and the ferrite rod aerial in the receiver. A metal box will act as a screen preventing reception of Radio 2. A metal box is only permissible if L2 is mounted outside and at least ${ }^{2} \mathrm{in}$. clear of the metal box, but this is rather getting away from the original idea of the Add-a-Verter.

Don't forget that ferrite rod aerials are highly directional and that the set and Add-a-Verter must be orientated for maximum signal strength of Radio 2.
Switch off the Add-a-Verter when normal reception is required on the medium waves as otherwise interfering beat notes will be heard on stations.
Transistors type 2 N 2926 with orange, yellow and green coding were all tried and found satisfactory but difficulty may be experienced if the lower gain types, brown or red, are used.

## BACK NUMBERS

We regret that the back numbers department of P.W. has now closed and consequently we are unable to supply these, Requests for specific back issues can usually be included in our 'CQ' section; there is no charge for this but it is a service between readers and P.W. are unable to meet any of these requests.


PART 2

NOTE:-In the Components List in Part 1 of the 24 Hour Digital Clock the reference number of the monolithic valtage regulator should have been shown as MVR5V and not MRV5V. The resistor marked R3 in Fig. 1 should have been shown as R4.

## CONSTRUCTION

Construction commences with the assembly of four decade counter boards for the display module. First of all four holes should be drilled at the extreme corners of the boards in order to take the 6BA studding with which the display module is finally assembled, plus another below the edge connector, Fig. 7.


Each board has the component values and locations clearly marked together with the relevant notation for the edge connections. The resistor marked Rl on each of the P.C. boards corresponds to the Rl's on the circuit diagram and has a value of $16 \mathrm{k} \Omega$ for the recommended tube. Tinned copper wire should be used for all the links shown on the boards. The two types of integrated circuits should now be mounted on the boards paying especial attention that the indentations on the i.c.'s align with the marks on the P.C. board.

The numicator tube is now mounted with its leads straddling the board and viewing direction to the front. The pin connections for the tube are shown in Fig. 3 and the positions for these connections are clearly shown on the board itself.

All four boards are assembled in exactly the same way. Three of the four boards must now be slightly modified as shown in Fig. 6 (boards 2, 3 and 4). This modification is made to separate the two zero reset connections in order to be able to modify the counting sequence as described in the circuit operation.

The boards are now assembled on the studding as shown in Fig. 7 and spaced using extra 6BA

Fig. 7. Assembly and interconnecting wiring of the display module comprising the four decade counter boards.


Fig. B. Constructional details and wiring of the countdown board. Care must be taken to ensure that the copper strips are cleared from around the two fixing down brackets.

Fig. 9. Layout of the power supply components and two reset switches on the back panel. The voltage regulator must be bolted directly on the back panel which acts as a heat sink.

nuts. Mounting feet are fitted to the P.C. boardsthese can be seen in the photographs.

The interconnections at the rear of the boards are now made as shown in Fig. 7. Connections may be made with tinned copper wire through the holes in the edge connectors. Flying leads are connected for the input, +5 V , chassis, and h.t. voltage. This completes the display module.

The countdown board is now assembled on a piece of $8 \cdot 1 \mathrm{in}$. matrix stripboard $3^{3}{ }_{4} \mathrm{in}$. long by $1{ }^{1} 4 \mathrm{in}$. wide. The layout is exactly as shown in the circuit diagram, the only cuts to be made to the board are two complete lines along the length of the board between the rows of i.c. pins, Fig. 8.

All the chassis connections of the i.c.'s are commoned as are the 5 V pins and flying leads added for connection to the power supply. Flying leads are also connected for the input, output and setting switches. This completes the countdown module.

The main chassis on which the power supply and other modules are mounted is simply an aluminium panel (18 s.w.g.) with an $L$ cross section. The rear panel carries the setting switches and power supply while other modules are bolted to the base. The transformer is bolted to the rear panel as is the monolithic voltage regulator which uses the rear panel as its heat sink, Fig. 9.
The two large capacitors are also fixed to the rear panel with vertical fixing clips. The remaining power supply components may be soldered to the tags of the components already mentioned and the four outlets taken to a tag panel to permit easy connection to the other two modules.

The countdown board and display module are now bolted on to the basic chassis using angle


Dimensions and drilling details for the case and chassis which can be obtained as a kit. Note that the flanges on the base plate are uppermost. The backplate is $\frac{3}{8} \frac{3}{8}$. shorter than the front plate thus leaving a slot along the back to provide essential ventilation. The metalwork shown in the various photographs of the Digital Clock was that used by the author in the prototype.
brackets and all the flying leads connected to the power supply and setting switches on the rear panel.

## SETTING UP

The initial setting of the clock to a particular time is carried out as follows:

SWl is depressed to allow the clock to run at PRESENT SETTING SYSTEM


## NEW SETTING SYSTEM



1500 times normal speed and released at a displayed time at least half an hour before the required setting. SW2 is now depressed which allows the clock to run at a speed 30 times greater than normal. The button is released at the desired time setting. The clock will now run to an accuracy determined only by the accuracy at which the mains frequency is maintained. Resetting after a power cut or accidental disconnection may be carried out as described above.

In some cases and with some types of switches, contact bounce can cause overshoot on the initial setting. It is possible to use extra integrated circuits together with changeover switches to provide electrical rather than mechanical control of the initial setting. However, due to the fact that the clock will only infrequently need setting, the extra circuit complexity and expense was not considered justified. A circuit to overcome contact bounce is shown in Fig. 10.


# semiconductor <br> dot <br>  



COMPILED FOR PRACTICAL WIRELESS BY D.A. LONGLAND, B.Sc., C. Eng., M.I.E.R.E. (ELECTROVALUE)

THIS SUPPLEMENT has been designed to present data to readers in a basic way. The technical details, parameters and capabilities of the more common types are given. The tables are broken down into categories by usage to simplify selection but of course many transistors could be included in several different headings. This has not been done as it would necessitate a considerable degree of duplication.

Transistors are used for many purposes, most of which may be reduced into the broad classification of increasing the power or level of a signal. This is effected by introducing the signal into the base-emitter (or gate-source circuit if a field effect transistor, f.e.t.) and thus causing an amplified variation of the signal in the collectoremitter circuit (or source drain circuit if anf.e.t.).
The important characteristics required of a transistor are determined by its application, that is, whether used for amplification of small a.c. or d.c. signals, for switching current or power amplification. Other characteristics may be important when a transistor is used near one of its limits, for example, high frequency, low noise, saturation, high power, very low current, or fast switching times.

## Transistor characteristics

An ideal transistor will have a high transition frequency (fT) and low feedback capacitance (for good high frequency response), stable characteristics, low noise level, low saturation voltage, fast rise and fall times (for low loss switching characteristics), high thermal conductivity (for high power handling capacity), low leakage (for ability to operate at very low currents), high gain (to reduce the number of stages for a given requirement) and low temperature effects (to reduce dependence on operating temperature and power dissipation).
It would be rare to find many extreme characteristics in one transistor for the technology used in producing an extreme in one characteristic may be incompatible with that used in producing an extreme in another. Hence, if a small signal, high frequency amplifier transistor were chosen it would have a medium gain at low currents, low feedback capacitance and high fT, therefore it would be made on a small semiconductor chip. One would not therefore expect the same transistor to handle a high voltage, high power, or have a low saturation voltage at high current. This illustrates the need for com-
promise and in turn accounts for the large number of transistors available, many having their own individual collections of characteristics suited to particular applications.
In choosing a transistor, one has to start by determining the category in which the major characteristics fall, e.g. r.f. or a.f. amplification, switching, low or high power, high or low voltage, PNP or NPN. After choosing a range of transistors, the secondary characteristics may then be chosen.

1. In high frequency application, having chosen one with an fT considerably higher than the operating frequency (if possible) and having a low feedback capacitance one would then seek a type having suitable voltage rating, power, dissipation, current or power gain, and if for switching, rise and fall time and saturation voltage.
2. For low frequency application, the fT is relatively unimportant though with a high gain transistor, a transition frequency rather higher than the product of the current gain and highest frequency of operation should be sought: thus an amplifier using a transistor with an fT of 1 MHz and a current gain of 100 would be expected to fall in frequency response at 10 kHz . A suitable voltage and power rating would then be chosen, keeping in mind any extremely low current or low saturation voltage characteristic required: thus an amplifier to operate at 1 mA should have a collector-base leakage current rather less than 1 mA divided by the desired maximum current gain of the transistor in order that the transistor will not saturate. A high gain transistor will give a higher input resistance than a low gain type, thus the high gain type would take less to drive power, but may cause a limitation in the frequency response.
3. For switching application it is desirable that the transistor conduct heavily with little voltage drop (i.e. low saturation voltage) when driven. For high speed switching the response time (i.e. rise, fall and storage) is important since the power dissipated in a slow transistor will be relatively high and may limit the power or frequency to be switched.
4. Other characteristics of a particular nature
may be sought, e.g. a given outline for incorporation into existing equipment, or ease of heat sinking (to reduce temperature rise if high power is to be dissipated), or high reverse base voltage for certain switching purposes, etc., etc. All must be considered in the design of equipment and in the selection of a suitable or substitute transistor.

The following tables are designed to assist in selection and substitution-it will be found that perhaps many transistors will serve one particular and non-demanding application whereas for another application only one or two will be found suitable. One factor to be considered is cost as it will be observed that a transistor which incorporates several special or extreme characteristics can be expensive. Good, common transistors which are relatively cheap and which will fulfil many needs are indicated by a dagger.

## Dissipation in power transistors

Under the High Current Audio headings, the power dissipation is given in terms of thermal conductivity between junction and case in $\mathrm{mW} /{ }^{\circ} \mathbf{C}$. There is some difficulty in expressing dissipation concisely as manufacturers are inclined to give dissipation ratings at different temperatures. It is felt that the thermal conductivity gives a more useful guide in the comparison between transistors.

In order to calculate the dissipation at a particular case temperature one must multiply the excess of maximum junction temperature above case temperature by the thermal conductivity. In some cases the manufacturer does not declare the dissipation at $25^{\circ} \mathrm{C}$, therefore one should not assume that it would be safe to dissipate the maximum power as calculated above at $25^{\circ} \mathrm{C}$ case temperature. One may assume it safe to operate at a case temperature midway between $25^{\circ} \mathrm{C}$ and maximum junction temperature, this covering most practical requirements.

As a general guide, the maximum junction temperatures (which should not be exceeded by $20^{\circ} \mathrm{C}$ ) are: Germanium (metal case) $90^{\circ} \mathrm{C}$ : Silicon (metal case) $200^{\circ} \mathrm{C}$; Silicon (plastic case) $150^{\circ} \mathrm{C}$ for types $2 \mathrm{~N}-, \mathrm{MJE}-, \mathrm{TIP}$ - and $125^{\circ} \mathrm{C}$ for types BD-.

| Abbreviations |  | Ib | Base current. |
| :---: | :---: | :---: | :---: |
|  |  | Icbo | Collector-base current with emitter |
| t | Typical. |  | open-circuit. |
| ax | Maximum. | NF | Noise factor. |
| min | Minimum. | PG | Power gain. |
| Cob | Maximum common-base output capacitance. | Vcbo | Collector-base voltage with emitter open-circuit. |
| f | fhfb, Frequency at which the common- | Vce | Collector-emitter voltage. |
| fT | Frequency at which the common- | Vce(sat) | Collector-emitter voltage for saturated (fully conducting) operation. |
|  | emitter current gain is unity. | Veb | Base-emitter reverse voltage. |
| $\begin{aligned} & \text { hFE } \\ & \text { hfe } \end{aligned}$ | Large signal current gain. Small signal current gain. | * | Under Ic/Ib. These refer to Ic with the |

## GERMANUM TRANSISTORS

## R.F. Applications, PNP

| Type | $\begin{aligned} & \mathbf{V}_{\text {cbo }} \\ & \text { max } \end{aligned}$ | $\begin{aligned} & V_{\mathrm{ce}} \\ & \max \end{aligned}$ | $\mathbf{V}_{\mathbf{e b}}$ $\max$ | $\begin{gathered} I_{c} \\ \max _{\max } \end{gathered}$ | $P_{\text {tot }}$ max | $h_{\text {FE }}^{h_{f e}}$ | $\underset{\mathbf{m A}}{\mathbf{I}_{\boldsymbol{c}}}$ | $\underset{\mu \mathrm{A}}{\mathbf{I}_{\text {cbo }}}$ | $\begin{aligned} & \mathbf{v}_{\text {ce }} \\ & \text { (sat) } \end{aligned}$ | $I_{c} / I_{b}$ mA | $\mathrm{HT}_{\mathrm{M}}$ | $\underset{\text { pF }}{\mathbf{C F}_{\text {ob }}}$ | $\begin{aligned} & \mathrm{NF} \\ & \mathrm{~dB} \\ & \max \end{aligned}$ | $\begin{aligned} & P G \\ & d B \end{aligned}$ | MHz | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| NKT613F | 40 | 40 | 1 | 10 | 80 | 40 min | 1 | 5 |  |  | $75 t$ | 2 |  | 14 | 100 | T07 |
| AF114 | 20 | 20 |  | 10 | 60 | -130t | 1 | 8 |  |  | $75 t$ | $2 \cdot 5$ |  | $12 \cdot 5$ | 100 | T07 |
| AF115 | 20 | 20 |  | 10 | 60 | -130t | 1 | 8 |  |  | $75 t$ |  |  | 10 | 100 | T07 |
| AF116 | 20 | 20 |  | 10 | 60 | -130t | 1 | 8 |  |  | $75 t$ |  |  | 19 | 10.7 | T07 |
| AF117 | 20 | 20 |  | 10 | 60 | -130t | 1 | 8 |  |  | 75t |  |  | 40 | 0.45 | T07 |
| AF118 | 70 | 70 | 1 | 30 | 375 | -180t | 1 |  |  |  | 175 | $2 \cdot 3$ |  |  |  | T07 |
| AF124 | 20 | 20 |  | 10 | 60 | *130t | 1 | 8 |  |  | $75 t$ | $2 \cdot 5$ |  | $12 \cdot 5$ | 100 | T072b |
| AF125 | 20 | 20 |  | 10 | 60 | *130t | 1 | 8 |  |  | $75 t$ |  |  | 10 | 100 | TO72b |
| AF126 | 20 | 20 |  | 10 | 60 | *130t | 1 | 8 |  |  | $75 t$ |  |  | 19 | 10.7 | TO72b |
| AF127 $\dagger$ | 20 | 20 |  | 10 | 60 | *130t | 1 | 8 |  |  | $75 t$ |  |  | 40 | 0.45 | TO72b |
| AF139 | 20 | 15 | 0.3 | 10 | 60 | 50 t | 1 | 8 |  |  | 5001 |  |  | 10 | 800 | T072a |
| AF201 | 25 |  | 0.3 0.3 | 10 | 225 | 50 t | 1 | 10 |  |  |  |  |  |  |  | TO72b |
| AF239 $\dagger$ | 20 | 15 | 0.3 | 10 | 60 | 504 | 1 | 8 |  |  | 700 |  |  | 14 | 800 | T072a |
| OC44 | 15 |  | 12 | 10 | 75 | 40-225 | 2 | 2 |  |  | $7 \cdot 51$ |  |  |  |  | TO1,RO9 |
| OC45 | 15 |  | 12 | 10 | 75 | 25-125 | 2 | 2 |  |  | 97 |  |  |  |  | T01,ROS |

## Low Noise, PNP



## Small Signal Audio, PNP

| Tуре | $V_{\text {ebo }}$ max | $V_{\text {ce }}$ max | $\begin{aligned} & V_{\text {ab }} \\ & \max \end{aligned}$ | 1. mA max | Prot max | $\begin{aligned} & \text { hfe } \\ & \text { thfe } \end{aligned}$ | $I_{c}$ mA | $\boldsymbol{I}_{\mu \mathrm{A}}$ | $\begin{aligned} & V_{c e} \\ & (\operatorname{set}) \end{aligned}$ | $\mathrm{I}_{\mathrm{c}} / \mathrm{l}_{\mathrm{b}}$ mA | $\mathrm{F}_{\mathbf{T}}$ | $\begin{gathered} \mathbf{C}_{\text {ob }} \\ \mathbf{p F} \end{gathered}$ | NF dB $\max$ | $\begin{aligned} & \text { PG } \mathbf{M H z} \\ & \mathbf{d B} \\ & \text { @ } \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| NKT213 | 32 | 32 | 10 | 250 | 200 | 50-130 | 1 | 10 | $0 \cdot 2$ | 25/2.5 | 0.97 | 60 |  |  | TO1 |
| NKT214 $\dagger$ | 32 | 32 | 10 | 250 | 200 | -30-75 | 1 | 10 | 0.2 | 25/2.5 | 0.97 | 60 | 10 |  | TO1 |
| NKT274 $\dagger$ | 15 | 15 | 5 | 250 | 200 | * 85-250 | 1 | 10 |  |  | 1 |  | 15 |  | TO1 |
| NKT275 | 15 | 15 | 5 | 250 | 200 | -30-90 | 1 | 10 |  |  | 1 |  | 10 |  | TO1 |
| OC71 | 30 | 20 | 10 | 10 | 125 | 30-75 | 3 | 13 |  |  | 0.3f |  | 16 |  |  |

## Medium Current Audio, Small Output Types, PNP

| Tуре | Vebo max | $\begin{aligned} & \mathbf{V}_{\mathrm{ce}} \\ & \max \end{aligned}$ | $\begin{aligned} & \mathbf{V}_{\text {eb }} \\ & \text { max } \end{aligned}$ | $\mathrm{I}_{\mathrm{E}}$ mA max | $P_{\text {tot }}$ max | $\begin{gathered} \mathbf{h}_{\text {FE }} \\ \boldsymbol{H}_{f 0} \end{gathered}$ | $I_{c}$ mA | Itbo $\mu \mathrm{A}$ | $\begin{aligned} & V_{e e} \\ & (s a t) \end{aligned}$ | $\mathrm{Ic} / \mathrm{I}_{\mathrm{b}}$ mA | $\stackrel{\mathrm{ft}_{\mathrm{Mz}}}{ }$ | Cob pF | NF $d B$ max | $\begin{aligned} & \text { PG MHz } \\ & \text { dB } \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AC128 $\dagger$ | 32 | 16 | 10 | 1000 | 220 | 40-110 | 300 | 10 |  |  | $1 \cdot 5$ | 100 |  |  | T01 |
| AC142H | 50 | 30 | 10 | 1200 | 220 | 40-110 | 400 | 14 |  |  | $1 \cdot 2$ |  |  |  | TO1 |
| AC142HK | 50 | 30 | 10 | 1200 | 860 | 40-110 | 400 | 14 |  |  | $1 \cdot 2$ |  |  |  | $\times 9$ |
| AC153 | 32 | 18 | 10 | 2000 | 220 | 50-250 | 300 | 200 |  |  | $1 \cdot 5$ |  |  |  | TO1 |
| AC153K $\dagger$ | 32 | 18 | 10 | 2000 | 1000 | 50-250 | 300 | 200 |  |  | $1 \cdot 5$ |  |  |  | X9a |
| NKT211 $\dagger$ | 32 | 30 | 10 | 1000 | 200 | 20-150 | 300 | 10 | $0 \cdot 5$ | $300 / 30$ | 0.9f | 60 |  |  | TO1 |
| NKT212 | 32 | 32 | 10 | 500 | 200 | 50-150 | 50 | 10 | $0 \cdot 15$ | 25/2-5 | 0.97 |  |  |  | T01 |
| NKT217 | 60 | 40 | 10 | 500 | 200 | 50-150 | 25 | 10 | $0 \cdot 5$ | 300/6 | 0.97 | 60 |  |  | TO1 |
| NKT271 | 15 | 15 | 5 | 500 | 200 | 50-250 | 50 | 10 |  |  | 1 f |  |  |  | T01 |
| OC72 | 32 | 32 | 10 | 250 | 125 | 30-90 | 80 | 15 | 0.25 | 125/12-5 | 1 |  |  |  | TO1,RO: |
| OC81 | 32 | 32 | 10 | 500 | 240 | 50-250 | 50 | 10 |  |  | 1 f |  |  |  | TO1 |
| ${ }^{\circ} \mathrm{C} 83$ | 32 | 20 | 3 | 500 | 220 | 50-280 | 50 | 100 | 0.5 | 300/9 | 0.85 |  |  |  | TO1 |
| OC84t | 32 | 20 | 10 | 500 | 220 | 60-200 | 50 | 100 | 0.5 | 300/9 | 0.85 |  |  |  | TO1 |

## Medium Current Audio, Small Output Types, NPN

| Type | $V_{\text {cbo }}$ max | $\begin{aligned} & V_{\text {ce }} \\ & \max \end{aligned}$ | $V_{\text {eb }}$ max | $\max _{\max _{\mathbf{m}}}$ | $P_{\text {tot }}$ $\max$ | $\underset{{ }_{*}^{h_{\text {Fe }}}}{h_{\text {FE }}}$ | $\begin{aligned} & \mathbf{I}_{\boldsymbol{c}} \\ & \mathrm{mA} \end{aligned}$ | $\underset{\mu \mathrm{A}}{\mathbf{I}_{\mathrm{Ebo}}}$ | $\begin{gathered} V_{c e} \\ \text { (sat) } \end{gathered}$ | $\begin{aligned} & \mathbf{I}_{\mathrm{c}} / \mathrm{I}_{\mathrm{b}} \\ & \mathrm{~mA} \end{aligned}$ | ${ }_{\mathrm{M}}^{\mathrm{M}} \mathrm{~Hz}_{2}$ | $\underset{\mathrm{pF}}{\mathbf{C F}_{\text {ob }}}$ | NF <br> dB <br> max | $\begin{aligned} & \text { PG } \mathbf{M H z} \\ & \mathbf{d B} \\ & \\ & \hline \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AC127 $\dagger$ | 32 | 32 | 10 | 500 | 200 | 25-143 | 500 | 10 | 1 | 500/50 | $1 \cdot 5$ |  |  |  | T01 |
| AC141H | 50 | 25 | 10 | 1200 | 220 | 40-110 | 400 | 14 | 0.55 | $400^{*}$ | 3 |  |  |  | T01 |
| AC141HK | 50 | 25 | 10 | 1200 | 860 | 40-110 | 400 | 14 | 0.55 | 400* | 3 |  |  |  | X9a |
| AC176 | 32 | 20 | 5 | 300 | 220 | 52-180 | 500 | 100 | 0.6 | 500/10 | 1 |  |  |  | TO1 |
| AC176K $\dagger$ | 32 | 20 | 5 | 300 | 1000 | 52-180 | 500 | 100 | 0.6 | 500/10 | 1 |  |  |  | X9a |
| NKT713 | 30 | 30 | 15 | 500 | 150 | 50-150 | 50 | 15 | 0.15 | 50/5 | $1 f$ |  |  |  | TO1 |
| NKT773 | 15 | 15 | 5 | 300 | 150 | 50-150 | 50 | 15 |  |  |  |  |  |  | TO1 |

High Current Audio, Power Output Types, PNP

| Type | $\begin{aligned} & \mathbf{V}_{\text {cbo }} \\ & \max \end{aligned}$ | $\begin{aligned} & V_{c e^{\circ}} \\ & \max \end{aligned}$ | $V_{e b}$ $\max$ | $\begin{gathered} \mathbf{J}_{\mathbf{c}} \\ \text { axx } \end{gathered}$ | Pto: max | ${ }_{*}^{h_{\mathrm{hf}}}$ | $\begin{aligned} & \mathbf{I}_{\mathbf{c}} \\ & \mathbf{m A} \end{aligned}$ | $\underset{\mu \mathrm{A}}{\mathbf{I}_{\mathrm{A}}}$ | $\begin{array}{\|c} v_{\text {ce }} \\ \text { (sat) } \end{array}$ | $I_{c} / I_{b}$ mA | $\mathbf{M}^{\mathbf{f}} \mathrm{Hz}^{2}$ | $\underset{\mathbf{p F}}{\mathbf{C F}_{\text {ob }}}$ | NF dB max | $\begin{aligned} & \text { PG } \mathbf{~ M H z ~} \\ & \mathbf{d B} @ \\ & \text { @ } \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AD140 | 55 |  | 10 | 3A | 666 | 30-100 | 1000 | 100 |  |  |  |  |  |  | T03 |
| AD142 | 80 | 50 | 10 | 10A | 666 | 30-170 | 1000 | 5000 | 0.3 | 5000/250 | 0.45 |  |  |  | T03 |
| AD149 $\dagger$ | 50 | 30 | 20 | $3 \cdot 5 \mathrm{~A}$ | 500 | 30-100 | 1000 | 350 |  |  | 0.5 |  |  |  | TO3 |
| AD162 $\dagger$ | 32 | 20 | 10 | 1A | 222 | 80-320 | 500 | 40 |  |  | 1 |  |  |  | MD17c |
| AL102 | 130 | 60 | 2 | 6A | 666 | 40-250 | 1000 | 1000 | 0.5 | 5000/250 | 4 |  |  |  | TO3 |
| NKT403 $\dagger$ | 80 | 32 | 40 | 10A | 770 | 50-150 | 1000 | 150 | 0.42 | 1000/100 | 0.35 0.35 |  |  |  | TO3 |
| NKT405 | 60 | 32 | 20 | 5A | 770 | 100-200 | 1000 | 150 | 0.42 | 1000/100 | 0.35 |  |  |  | TO3 |
| OC25 | 40 | 40 | 10 | 4A | 500 | 15-80 | 1000 | 100 |  |  |  |  |  |  | TO3 |
| OC28 | 80 | 60 | 40 | 8A | 666 | 20-55 | 1000 | 100 | $1 \cdot 6$ | ${ }^{6000}{ }^{*}$ | 0.25 f |  |  |  | TO3 |
| OC29 $\dagger$ | 60 | 48 | 20 | 8A | ${ }_{666}^{666}$ | 45-130 | 1000 | 100 | 1.6 | $6000 *$ 6000 | $0.25 f$ $0.25 f$ 0.25 |  |  |  | TO3 |
| OC35 | 60 | 48 | 20 | 8A | 666 666 | ${ }_{\text {25-75 }}$ | 1000 | 100 | 1.4 1.6 | $6000^{*}$ 6000 | $0.25 f$ $0.25 f$ |  |  |  |  |
| OC36 2N2147 | 80 75 | 60 50 | 40 1.5 | 8A | 666 666 | $30-110$ 100 min | 1000 1000 | 100 1000 |  |  |  |  |  |  | TO3 |
| 2N2147 | 75 | 50 | 1.5 | 5A | 666 | 100 min | 1000 | 1000 | 0.6 | 5000/250 | 3 |  |  |  |  |

High Current Audio, Power Output Types, NPN

| AD161t | 32 | 20 | 10 | 14 | 222 | $80-320$ | 500 | 50 |  |  | 1 |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| MD17c |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

See note on power dissipation

## Medium Current Switching, PNP

| Type | $\begin{aligned} & \mathbf{V}_{\text {cbo }} \\ & \text { max } \end{aligned}$ | $\begin{aligned} & \mathbf{V}_{\mathrm{cq}} \\ & \max \end{aligned}$ | $V_{e b}$ max | ${ }_{\mathrm{m}}^{\mathrm{I}}{ }_{\mathrm{m}}$ <br> max | $P_{\text {tot }}$ max | $h_{\text {FE }}$ <br> ${ }^{*} h_{f e}$ | $I_{c}$ $m A$ | Icbo $\mu \mathrm{A}$ | $\begin{gathered} \mathbf{V}_{c e} \\ \text { (sat) } \end{gathered}$ | $I_{c} / I_{b}$ mA | $\frac{\mathrm{ft}}{\mathrm{MHz}}$ | $\mathbf{c}_{\mathbf{o b}}$ $\mathrm{pF}$ | $\begin{aligned} & \text { NF } \\ & \text { dB } \\ & \max \end{aligned}$ | PG MHz $d B$ <br> (a) | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ACY17 | 70 | 32 | 12 | 500 | 260 | 50-150 | 300 | 10 | $0 \cdot 3$ | 300/15 | 1 |  |  |  | TO5 |
| ACY18 | 50 | 30 | 12 | 500 | 260 | 40-120 | 300 | 10 | $0 \cdot 3$ | 300/15 |  |  |  |  | TO5 |
| ACY19 | 50 | 30 | 12 | 500 | 260 | 80-250 | 300 | 10 | $0 \cdot 3$ | 300/15 | $1 \cdot 5$ |  |  |  | T05 |
| ACY20 | 40 | 20 | 12 | 500 | 260 | 50-145 | 50 | 10 | $0 \cdot 2$ | 50/1•3 | 1 |  |  |  | TO5 |
| ACY21 | 40 | 20 | 12 | 500 | 260 | 90-250 | 50 | 10 | $0 \cdot 2$ | 50/1•3 | 1 |  |  |  | TO5 |
| ACY22 | 20 | 15 | 12 | 500 | 260 | 30-300 | 300 | 10 | $0 \cdot 3$ | 300/15 | 0.8 |  |  |  | TO5 |
| ACY39 | 110 | 40 | 12 | 500 | 260 | 50-150 | 300 | 10 | $0 \cdot 3$ | 300/15 | 1 |  |  |  | TO5 |
| 2N1303 | 30 | 25 | 25 | 200 | 150 | 20 min | 10 | 6 | 0.2 | 10/0.5 | 5 | 20 |  |  | TO5 |
| 2N1305 $\dagger$ | 30 | 20 | 25 | 200 | 150 | 40-200 | 10 | 6 | $0 \cdot 2$ | 10/0.25 | 10 | 20 |  |  | TO5 |
| 2N1307 | 30 | 15 | 25 | 200 | 150 | 60-300 | 10 | 6 | $0 \cdot 2$ | 10/0-17 | 15 | 20 |  |  | TO5 |
| 2N1309 | 30 | 15 | 25 | 200 | 150 | 80 min | 10 | 6 | $0 \cdot 2$ | 10/0.13 | 20 | 20 |  |  | TO5 |
| ASY26 | 30 | 15 | 20 | 300 | 150 | 30-80 | 20 | 7 | $0 \cdot 25$ | 50/2 | 8 | 15 |  |  | TO5 |
| ASY27 | 25 | 15 | 20 | 300 | 150 | 50-150 | 20 | 7 | $0 \cdot 25$ | 50/1 - 55 | 14 | 16 |  |  | TO5 |

## Medium Current Switching, NPN

| Type | $\begin{aligned} & \mathbf{V}_{\text {cbo }} \\ & \max \end{aligned}$ | $V_{c}$ max | $\begin{aligned} & \mathbf{V}_{\text {eb }} \\ & \max \end{aligned}$ | $\max _{\mathrm{ma}_{c}}$ | $\mathbf{P}_{\text {tot }}$ $\max$ | $h_{h_{f e}}^{h_{f E}}$ | $\begin{aligned} & \mathbf{I}_{c} \\ & \mathbf{m A} \end{aligned}$ | $\begin{aligned} & \mathbf{I f b e b}_{\mu \mathbf{A}} \end{aligned}$ | $\begin{gathered} \mathbf{V}_{\text {ce }} \\ \text { (sat) } \end{gathered}$ | $\begin{aligned} & \mathbf{I}_{c} / I_{b} \\ & m A \end{aligned}$ | $\stackrel{\mathbf{f} \mathbf{T}}{\mathbf{M H z}}$ | $\underset{\mathrm{pF}}{\mathrm{C}_{\mathrm{ob}}}$ | $\begin{gathered} \mathrm{NF} \\ \mathrm{~dB} \\ \max \end{gathered}$ | $$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASY28 | 30 | 15 | 20 | 200 | 150 | 30-80 | 20 | 35 | 0.25 | 50/2 | 14 | 16 |  |  | TO5 |
| ASY29 | 25 | 15 | 20 | 200 | 150 | 50-150 | 20 | 35 | 0.25 | 50/1.25 | 20 | 16 |  |  | T05 |
| 2N1302 | 25 | 25 | 25 | 200 | 150 | 20 min . | 10 | 6 | 0.2 | 10/0.5 | 10 | 20 |  |  | TO5 |
| 2N1304 $\dagger$ | 25 | 20 | 25 | 200 | 150 | 40-200 | 10 | 6 | 0.2 | 10/0.25 | 15 | 20 |  |  | T05 |
| 2N1306 | 25 | 15 | 25 | 200 | 150 150 | $60-300$ 80 min . | 10 10 | 6 | 0.2 0.2 | 10/0.17 $10 / 0 \cdot 13$ | 20 30 | 20 |  |  | T05 |
| 2N1308 | 25 | 15 | 25 | 200 | 150 | 80 min . | 10 | 6 | 0.2 | 10/0.13 | 30 | 20 |  |  |  |

## High Current Switching, PNP

| Type | $V_{\text {cbo }}$ $\max$ | $V_{c e}$ max | $\begin{array}{\|c} \mathbf{V}_{\text {eb }} \\ \mathrm{max} \\ \hline \end{array}$ | $\operatorname{If}_{\max }$ | Ptot max | $\begin{gathered} h_{\text {FE }} \\ { }^{h_{f E}} \end{gathered}$ | $\begin{gathered} \mathbf{I}_{c} \\ \mathbf{m A} \end{gathered}$ | $\underset{\mu \mathrm{A}}{\mathrm{I}_{\mathrm{tbo}}}$ | $\begin{aligned} & V \\ & \text { csa } \end{aligned}$ | $\mathrm{I}_{\mathrm{c}} / \mathrm{l}_{\mathrm{b}}$ mA | $\stackrel{\mathrm{fT}}{\mathrm{M} \mathrm{~Hz}}$ | Cob pF | NF <br> dB <br> max | $\begin{aligned} & P G \\ & d B \end{aligned}$ | ${ }^{\mathbf{M H z}}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AU111 | 320 |  | 2 | 10A |  | 15-80 | 6000 |  | 0.5 | 5000/250 | 2 |  |  |  |  | TO3 |

The term switching above does not imply any discontinuity in the characteristics. Such transistors may be used as amplifiers, as may amplifier transistors be used for switching duty if their characteristics are suitable.


## R.F. Applications, NPN

| Tуре | $\mathbf{V}_{\text {ebo }}$ max | $\begin{aligned} & \mathbf{V}_{\text {ce }} \\ & \max \end{aligned}$ | $\mathbf{V}_{\mathrm{eb}}$ max | $I_{c}$ mA max | Ptot max | hfe <br> *hfo | $t_{c}$ mA | $I_{\text {ebo }}$ $\mu \mathrm{A}$ | $\begin{aligned} & V_{c t} \\ & \text { (sat) } \end{aligned}$ | $\mathrm{I}_{\mathrm{c}} / \mathrm{I}_{\mathrm{b}}$ mA | $\stackrel{\text { fT }}{\mathrm{MHz}}$ | $\begin{gathered} \text { Cob }_{\text {ob }} \\ \text { pF } \end{gathered}$ | NF $d B$ max | PG dB | MHz | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BF115 | 50 | 30 | 5 | 30 | 145 | 100 t | 3 |  |  |  | 2304 |  | 4 |  |  | T072b |
| BF167 | 40 | 30 | 4 | 25 | 130 | 50 t | 4 |  |  |  | 3501 |  | $3 \cdot 5$ |  |  | TO72b |
| 8F173 $\dagger$ | 40 | 25 | 4 | 25 | 260 | 87t | 7 |  |  |  | $550 t$ |  |  |  |  | T072b |
| 8F194t | 30 | 20 | 5 | 30 | 220 | 115t | 1 |  |  |  | 260t |  | 4 |  |  | MM10b |
| BF195 | 30 | 20 | 5 | 30 | 220 | 67t | 1 |  |  |  | 200t |  | 4 |  |  | MM10b |
| BF254t | 30 | 20 | 5 | 30 | 220 | 115t | 1 |  |  |  | $260 t$ |  | 4 |  |  | TO92z |
| BF255 | 30 | 20 | 5 | 30 | 220 | 67 t | 1 |  |  |  | $200 t$ |  | 4 |  |  | - TO92z |
| BFY90 | 30 | 15 | $2 \cdot 5$ | 50 | 200 | 60 t | 2 |  |  |  | 1000 | $1 \cdot 8$ | 5 | 224 | 500 | TO72a |
| 2N3663 | 30 | 12 | 3 | 25 | 200 | $75 t$ | 8 | 0.5 |  |  | 1250 t | $1 \cdot 7$ | $6 \cdot 5$ | 15 | 200 | TO98 |
| 2N4292t | 30 | 15 | 3 | 500 | 200 | 20 min | 3 | 0.5 |  |  | 570 t | $2 \cdot 2$ | 6 | $29 t$ | 100 | U29 |

Small Signal Audio, PNP

| Type | Vebo max | $\mathbf{V}_{\text {ce }}$ $\max$ | $\mathbf{V}_{e b}$ $\max$ |  | $\begin{aligned} & \text { Ptot }^{\text {max }} \end{aligned}$ | hfe <br> ${ }^{*} h_{f}$ | $\begin{aligned} & \mathbf{I}_{c} \\ & \mathrm{~mA} \end{aligned}$ | $t_{\text {cbo }}$ $\mu \mathrm{A}$ | $\begin{aligned} & \mathbf{V}_{\mathrm{ce}} \\ & (\mathrm{sat}) \end{aligned}$ | $\mathrm{t}_{\mathrm{c}} / \mathrm{I}_{\mathrm{b}}$ mA | fTHz | $\begin{gathered} \mathbf{C o b b}_{\text {ob }} \end{gathered}$ | NF $d B$ $\max$ | $\begin{gathered} \mathbf{P G} \text { NHz } \\ \mathbf{d B} \end{gathered}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BC157 $\dagger$ | 50 | 45 | 6 | 100 | 220 | *125-250 | 2 | 0.05 | 0.21 | 100/5 | $130 t$ |  |  |  | MM10a |
| BC158 | 30 | 20 | 5 | 100 | 220 | *125-500 | 2 | 0.1 | 0.24 | 100/5 | 130 t |  |  |  | MM10a |
| BC177 $\dagger$ | 45 | 45 | 5 | 100 | 300 | *125-260 | 2 | 0.05 | 0.21 | 100/5 | 130 t | 4 |  |  | TO18 |
| BC178 | 20 | 20 | 5 | 100 | 300 | *125-500 | 2 | $0 \cdot 1$ | $0.2 t$ | 100/5 | 130 t | 4 |  |  | T018 |
| BC186 | 40 | 25 | 5 | 100 | 300 | 40 min | 2 |  |  |  | 60 | 5 |  |  | T018 |
| BC212L | 60 | 50 | 5 | 100 | 360 | 60-300 | 2 | 0.015 | $0 \cdot 2$ | 10/0.5 | 200 | 10 |  |  | T092a |
| BC213L | 45 | 30 | 5 | 100 | 360 | 80-400 | 2 | 0.015 | $0 \cdot 2$ | 10/0. 5 | 200 | 10 |  |  | T092a |
| BC257t | 50 | 45 | 5 | 100 | 220 | *125-260 | 2 | 0.1 | 0.5 | 100/5 | 130t | 6 |  |  | T092a |
| BC258 | 30 | 25 | 5 | 100 | 220 | *125-500 | 2 | $0 \cdot 1$ | 0.5 | 100/5 | 130 t | 6 |  |  | T092a |
| BCY70 | 50 | 40 | 5 | 200 | 350 | 50 min . | 10 | 0.01 | 0.25 | 10/1 | 250 |  |  |  | T018 |
| BCY72 | 25 | 25 | 5 | 200 | 350 | 50 min . | 10 | $0 \cdot 1$ | 0.25 | 10/1 | 200 |  |  |  | T018 |
| ZTX500 | 25 | 25 | 5 | 500 | 300 | 50-300 | 10 | 0.2 | 0.35 | 50/5 | 150 | 6 |  |  | $\times 59$ |
| ZTX501 | 35 | 35 | 5 | 500 | 300 | 50-300 | 10 | $0 \cdot 2$ | 0.25 | 50/5 | 150 | 6 |  |  | $\times 59$ |
| ZTX502 | 35 | 35 | 5 | 500 | 300 | 100-300 | 10 | $0 \cdot 2$ | $0 \cdot 25$ | 50/5 | 150 | 6 |  |  | $\times 59$ |
| ZTX503 | 45 | 45 | 5 | 500 | 300 | 50-300 | 10 | $0 \cdot 2$ | 0.35 | 50/5 | 150 | 6 |  |  | $\times 59$ |
| ZTX504 | 70 | 70 | 5 | 500 | 300 | 50-300 | 10 | 0.2 | 0.35 | 50/5 | 150 | 6 |  |  | $\times 59$ |
| ZTX530 | 30 | 30 | 5 | 500 | 250 | 100-400 | $0 \cdot 1$ | 0.2 | $0 \cdot 7$ | 10/0.5 | 30 | 8 |  |  | $\times 59$ |
| ZTX531 | 45 | 45 | 5 | 500 | 250 | 40-120 | 0.01 | $0 \cdot 2$ | 0.7 | 10/0. 5 | 30 | 8 |  |  | $\times 59$ |
| 2N3906 | 40 | 40 | 5 | 200 | 310 | 100-300 | 10 | 0.05 | 0.25 | 10/1 | 250 | $4 \cdot 5$ |  |  | T092b |
| 2N4126 | 25 | 25 | 4 | 200 | 310 | 120-360 | 2 | 0.05 | 0.4 | 50/5 | 250 | $4 \cdot 5$ |  |  | TO92b |
| 2N4059 | 30 | 30 | 6 | 30 | 250 | 45-660 | 1 |  | 0.7 | 10/0. 5 |  |  |  |  | TO92a |
| 2N4060 | 30 | 30 | 6 | 30 | 250 | 45-185 | 1 |  | $0 \cdot 7$ | 10/0. 5 |  |  |  |  | TO92a |
| 2N4061 | 30 | 30 | 6 | 30 | 250 | 90-330 | 1 |  | 0.7 | 10/0. 5 |  |  |  |  | TO92a |
| 2N4062 2N4289 | 30 60 | 30 45 | 6 | 30 | 250 | 180-660 | 1 |  | 0.7 | 10/0.5 |  |  |  |  | T092a |
| 2N4289 | 60 | 45 | 7 | 500 | 250 | 160 min . | 1 | 0.01 | $0 \cdot 35$ | $1 / 0 \cdot 1$ | 170t | 8 |  |  | U29 |

## Small Signal, Audio, NPN

| Type | $V_{\text {cho }}$ $\max$ | $\begin{gathered} \mathbf{V}_{\mathbf{c e}} \\ \max \end{gathered}$ | $\begin{array}{\|l} \mathbf{V}_{\text {eb }} \\ \max \end{array}$ | $\begin{gathered} \mathbf{I}_{\mathbf{c}} \\ \max _{\max } \end{gathered}$ | $\begin{aligned} & P_{\text {tot }} \\ & \max \end{aligned}$ | ${ }_{{ }_{4} h_{f E}}$ | $\begin{aligned} & \mathbf{I}_{\mathbf{c}} \\ & \mathrm{mA} \end{aligned}$ | $\underset{\mu \mathrm{A}}{\mathrm{I}_{\text {cho }}}$ | $\begin{gathered} \mathbf{V}_{\mathbf{c e}} \\ \text { (sat) } \end{gathered}$ | $1 \mathrm{c} / \mathrm{I}_{\mathrm{b}}$ mA | $\frac{\mathbf{f H}_{\mathbf{T}}}{}$ | $\underset{\mathrm{pF}}{\mathrm{C}_{\mathrm{ob}}}$ | $\underset{\max }{\mathrm{NF}}$ | $\begin{aligned} & P G \\ & d B \end{aligned}$ |  | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BC107 $\dagger$ | 45 | 45 | 5 | 100 | 300 | *125-500 | 2 | 0.01 | 0.25 | 10/0.25 | 150 | $3 \cdot 7$ | 10 |  |  | T018 |
| BC108 | 20 | 20 | 5 | 100 | 300 | *125-900 | 2 | 0.01 | 0.25 | 10/0.5 | 150 | $3 \cdot 7$ | 10 |  |  | T018 |
| BC147 $\dagger$ | 50 | 45 | 6 | 200 | 220 | *125-500 | 2 | 0.02 | 0.25 | 10/0.5 | $300 t$ | $2 \cdot 5$ | 10 |  |  | MM10a |
| BC148 | 30 | 20 | 5 | 100 | 300 | *125-900 | 2 | 0.25 | 0.25 | 10/0.5 | $300 t$ | $2 \cdot 5$ | 10 |  |  | MM10a |
| BC167 $\dagger$ | 50 | 45 | 6 | 100 | 300 | *125-500 | 2 | 0.2 | 0.25 | 10/0. 5 | 3004 |  | 10 |  |  | TO92a |
| BC168 | 30 | 20 | 5 | 100 | 300 | *125-900 | 2 | 0.2 | 0.25 | 10/0.5 | 3004 |  | 10 |  |  | T092a |
| BC182L | 60 | 50 | 5 | 200 | 300 | 100-480 | 2 | 0.015 |  |  | 150 | 5 |  |  |  | TO92a |
| BC183L | 45 | 30 | 5 | 200 | 300 | 100-850 | 2 | 0.015 |  |  | 150 | 5 |  |  |  | TO92a |
| BC267 | 50 | 45 | 6 | 1000 | 375 | *125-500 | 2 | 0.015 | $0.22 t$ | 200/10 | 150 |  | 3 t |  |  | TO18 |
| BC268 | 30 | 20 | 6 | 1000 | 375 | *125-900 | 2 | 0.015 | 0.22 t | 200/10 | 150 |  | 3t |  |  | TO18 |
| ZTX300 | 25 | 25 | 5 | 500 | 300 | 50-300 | 10 | 0.2 | 0.35 | 50/5 | 150 | 6 |  |  |  | $\times 59$ |
| ZTX301 | 35 | 35 | 5 | 500 | 300 | 50-300 | 10 | 0.2 | 0.25 | 50/5 | 150 | 6 |  |  |  | $\times 59$ |
| ZTX302 | 35 | 35 | 5 | 500 | 300 | 100-300 | 10 | 0.2 | 0.25 | 50/5 | 150 | 6 |  |  |  | $\times 59$ |
| ZTX303 | 45 | 45 | 5 | 500 | 300 | 50-300 | 10 | 0.2 | 0.35 | 50/5 | 150 | 6 |  |  |  | $\times 59$ |
| ZTX304 | 70 | 70 | 5 | 500 | 300 | 50-300 | 10 | 0.2 | 0.35 | 50/5 | 150 | 6 |  |  |  | X59 |
| ZTX330 | 30 | 30 | 5 | 500 | 250 | 100-400 | 10 | 0.2 | 0.7 | 10/0.5 | 30 | 8 |  |  |  | $\times 59$ |
| ZTX331 | 45 | 45 | 5 | 500 | 250 | 40-120 | 10 | 0.2 | 0.7 | 10/0.5 | 30 | 8 |  |  |  | $\times 59$ |
| 2N2924 | 25 | 25 | 5 | 100 | 200 | *150-300 | 2 | 0.5 |  |  | 1201 | 12 | $2 \cdot 5 t$ |  |  | TO98 |
| 2N2925 | 25 | 25 | 5 | 100 | 200 | *235-470 |  | 0.5 |  |  | 120t | 12 | $2 \cdot 5 \mathrm{t}$ |  |  | TO98 |
| 2N2926B | 18 | 18 | 5 | 100 | 200 | *35-70 |  | 0.5 |  |  | 120t | 12 | $2 \cdot 5 t$ |  |  | T098 |
| 2N2926R | 18 | 18 |  | 100 | 200 | *55-110 |  | 0.5 |  |  | 120t | 12 | 2.5t |  |  | TO98 |
| 2N29260 | 18 | 18 | 5 | 100 | 200 | *90-180 | 2 | 0.5 |  |  | 120t | 12 | $2 \cdot 5 t$ |  |  | T098 |
| 2N2926 Y $\dagger$ | 18 | 18 | 5 | 100 | 200 | *150-300 | 2 | 0.5 |  |  | 120t | 12 | 2.5t |  |  | TO98 |
| 2N2926G $\dagger$ | 18 | 18 | 5 | 100 | 200 | *235-470 | 2 | 0.5 |  |  | 120t | 12 | $2 \cdot 5 t$ |  |  | TO98 |
| 2N3708 | 30 | 30 | 6 | 30 | 250 | 45-660 | 1 | 0.1 | 1 | 10/0. 5 |  |  |  |  |  | T092a |
| 2N3709 | 30 | 30 | 6 | 30 | 250 | 45-165 | 1 | $0 \cdot 1$ | 1 | 10/0.5 |  |  |  |  |  | T092a |
| 2N3710 | 30 | 30 |  | 30 | 250 | 90-330 | 1 | 0.1 | 1 | 10/0. 5 |  |  |  |  |  | T092a |
| 2N3711 | 30 | 30 | 6 | 30 | 250 | 180-660 | 1 | 0.1 | 1 | 10/0.5 |  |  |  |  |  | TO92a |
| 2N3904 | 60 | 40 | 5 | 100 | 310 | 100-300 | 10 | 0.1 | 0.2 | 10/1 | 200 t | 4 |  |  |  | TO92b |
| 2N4124 | 30 | 25 | 5 | 200 | 310 | 120-360 | 2 | 0.05 | $0 \cdot 3$ | 50/5 | 300 | 4 |  |  |  | T092b |
| 2N4286 $\dagger$ | 30 | 25 | 6 | 500 | 250 | 100 min | 0.1 | 0.05 |  |  | 2801 | 10 |  |  |  | ${ }_{\text {U29 }}{ }^{\text {298 }}$ |
| 2N5172 | 25 | 25 | 5 | 100 | 200 | *100-750 | 10 | $0 \cdot 1$ | 0.25 | -10/1 | 120t | 10 |  |  |  | TO98 |

## Low Noise, NPN

| Type | $\mathbf{V}_{\text {ebo }}$ max | $V_{c e}$ max | $\begin{gathered} \mathbf{V e b}_{\mathrm{max}} \end{gathered}$ | ${ }^{1}$ mA $\max$ | Ptot max | hfe ${ }^{*} h_{f e}$ | $\mathbf{I}_{c}$ $\mathrm{mA}$ | Icbo 1 A | $\begin{gathered} \mathbf{V}_{\mathrm{es}} \\ (\mathrm{sat}) \end{gathered}$ | $\\| \mathrm{l} / \mathrm{I}_{\mathrm{b}}$ mA | ${ }^{\mathbf{n}^{\mathbf{T}} \mathrm{Hz}}$ | $\underset{\text { pF }}{\mathbf{C F}_{\text {ob }}}$ | $\begin{aligned} & \mathrm{NF} \\ & \mathrm{~dB} \\ & \max \end{aligned}$ | $\begin{aligned} & \mathbf{P G} \mathbf{M H z} \\ & \mathbf{d B} \\ & \text { @ } \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BC109B $\dagger$ | 20 | 20 | 5 | 100 | 300 | *240-500 | 2 | 0.015 | 0.25 | 10/0.25 | 300t |  | 4 |  | T018 |
| BC109C $\dagger$ | 20 | 20 | 5 | 100 | 300 | * 450-900 | 2 | 0.015 | 0.25 | 10/0.25 | $300 t$ |  | 4 |  | T018 |
| BC149 $\dagger$ | 30 | 30 | 5 | 200 | 220 | *240-900 | 2 | 0.02 | 0.25 | 10/0.25 | $300 t$ |  | 4 |  | MM10a |
| BC169 $\dagger$ | 30 | 30 | 5 | 100 | 300 | *240-900 | 2 | 0.02 | 0.25 | 10/0-25 | $300 t$ |  | 4 |  | TO92a |
| BC184L | 45 | 30 | 5 | 200 | 300 | 250 min | 2 | 0.015 |  |  | 150 |  | 4 |  | TO92a |
| BC269 | 30 | 20 | 6 | 1000 | 375 | *240-900 | 2 | 0.015 | 0.22 t | 200/10 | 150 |  | 4 |  | T018 |
| 2N930 $\dagger$ | 45 | 45 | 5 | 60 | 300 | 100-300 | 0.01 | 0.01 |  |  | 80 t | 8 | 3 |  | T018 |
| 2N2483 | 60 | 60 | 6 | 50 | 360 | * 40-120 | 0.01 | 0.01 |  |  | 60 | 6 | 4 |  | T018 |
| 2N2484 | 60 | 60 | 6 | 50 | 350 | *100-500 | 0.01 | 0.01 |  |  | 60 | 6 | 3 |  | T018 |
| 2N3707 | 30 | 30 | 6 | 30 | 250 | *100-400 | 0.1 | 0.1 |  |  |  |  | 5 |  | T092a |
| 2N5088 | 35 | 30 | 3 | 50 | 310 | *300-900 | 0.1 | 0.05 |  |  | 50 | 4 | 3 |  | TO92b |

## Low Noise, PNP

| Type | $V_{\text {cbo }}$ max | $\underset{\max }{V_{c e}}$ | $V_{\text {max }}^{V_{a x}}$ | $\mathrm{I}_{\mathrm{c}}$ max | $P_{\text {tot }}$ max | $\underset{{ }_{*}^{h} h_{f e}}{ }$ | $\begin{aligned} & \mathbf{I}_{\mathbf{c}} \end{aligned}$ | $\begin{gathered} \mathbf{I}_{\mathrm{cbo}} \\ \mu \mathbf{A} \end{gathered}$ | $\begin{gathered} \mathbf{V}_{c o} \\ \text { (sat) } \end{gathered}$ | $1 \mathrm{c} / \mathrm{l}_{\mathrm{b}}$ mA |  | $\underset{\mathbf{p F}}{\mathbf{C F}_{\text {ob }}}$ | $\begin{gathered} \mathrm{NF} \\ \mathrm{~dB} \\ \max \end{gathered}$ | $\underset{\text { (a) }}{\mathbf{P G}}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BC153 | 40 | 40 | 5 | 100 | 200 | 50-135 | 0.1 | 0.002 | 0.25 | 10/0.5 | 40 | 6 | $2 \cdot 5$ |  | RO97a |
| BC154 $\dagger$ | 40 | 40 | 5 | 100 | 200 | 160-330 | 0.1 | 0.002 | 0.25 | 10/0.5 | 40 | 6 | $2 \cdot 5$ |  | R097a |
| BC159 $\dagger$ | 30 | 20 | 5 | 100 | 200 | *125-500 | 2 | 0.05 | $0 \cdot 3 t$ | 100/5 | 200t | 5 | 4 |  | MM10a |
| BC179 $\dagger$ | 20 | 20 | 5 | 100 | 300 | *125-900 | 2 | 0.05 | 0.34 | 100/5. | 2004 | 4 | 4 |  | TO18 |
| BC214L | 45 | 30 | 5 | 100 | 360 | 140-400 | 2 | 0.015 | 0.2 | 10/0.5 | 200 | 10 | 2 |  | T092a |
| BC259 $\dagger$ | 25 | 20 | 5 | 100 | 220 | 120-460 | 2 | 0.1 | 0.8 | 10/0.5 | 130 | - | 4 |  | TO92a |
| BCY71 | 45 | 45 | 5 | 200 | 350 | 100 min | 10 | 0.01 | 0.25 | 10/1 | 300 |  |  |  | T018 |
| 2N4058 | 30 | 30 | 6 | 30 | 250 | 100-400 | $0 \cdot 1$ | 0.1 | 0.7 | 10/1 |  |  | 5 |  | T092a |

Medium Current Audio, Small Output Types, NPN

| Type | Vcho max | $\mathbf{V}_{\text {c }}$ max | $\begin{aligned} & \mathbf{V}_{\text {eb }} \\ & \max \end{aligned}$ | $I_{c}$ mA max | $P_{\text {fot }}$ max | $\begin{gathered} \mathbf{h}_{\text {FE }} \\ \mathbf{H}_{\mathrm{fe}} \end{gathered}$ | $I_{c}$ mA | Icbo $\mu \mathrm{A}$ | $\begin{aligned} & V_{c o} \\ & \text { (sat) } \end{aligned}$ | $\begin{aligned} & I_{c} / I_{b} \\ & m A \end{aligned}$ | $\begin{gathered} \mathrm{f}_{\mathbf{T}} \\ \mathrm{MHz} \end{gathered}$ | $\begin{gathered} \mathbf{C}_{\text {ob }} \\ \mathrm{pF} \end{gathered}$ | $\begin{aligned} & \mathbf{N F} \\ & \mathbf{d B} \end{aligned}$ $\max$ | PG MHz dB <br> $\omega$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BC125 | 50 | 30 | 5 |  | 300 | 30 min | 150 | 0.05 | 0.5 | 150/15 | 40 | 12 |  |  | R097 |
| BC140-10 | 80 | 40 | 7 | 1000 | 750 | 20 min | 1000 | 0.1 | 1 | 1A/100 | 50 | 25 |  |  | TO5 |
| BC301 | 90 | 60 | 7 | 1000 | 850 | 40-240 | 150 | 0.02 | 0.5 | 150/15 | $120 t$ |  |  |  | TO5 |
| BFX84t | 100 | 60 | 6 | 1000 | 800 | 30 min | 150 | 0.05 | $0 \cdot 35$ | 150/15 | 50 |  |  |  | TO5 |
| BFX85 | 100 | 60 | 6 | 1000 | 800 | 70 min | 150 | 0.05 | 0.35 | -150/15 | 50 |  |  |  | TO5 |
| BFY50 | 80 | 35 | 6 | 1000 | 800 | 30 min | 150 | 0.05 | $0 \cdot 2$ | 150/15 | 60 |  |  |  | TO5 |
| BFY51 $\dagger$ | 60 | 30 | 6 | 1000 | 800 | 40 min | 150 | 0.05 | $0 \cdot 35$ | 150/15 | 50 |  |  |  | TO5 |
| BFY52 | 40 | 40 | 6 | 1000 | 800 | 60 min | 150 | 0.5 | $0 \cdot 35$ | 150/15 | 150 |  |  |  | TO5 |
| MC140 | 80 | 40 | 7 | 1000 | 800 | 40-300 | 150 | $0 \cdot 5$ | 0.7t | 1A/100 | 100 |  |  |  | X58b |
| 2N696 | 60 | 40 | 5 | 500 | 600 | 20-60 | 150 | 1 | $1 \cdot 5$ | 150/15 | 40 | 35 |  |  | TO5 |
| 2N697 | 60 | 40 | 5 | 500 | 600 | 40-120 | 150 | 1 | $1 \cdot 5$ | 150/15 | 50 | 35 |  |  | TO5 |
| 2N1613 | 75 |  | 7 | 500 | 800 | 40-120 | 150 | 0.01 | $1 \cdot 5$ | 150/15 | 60 | 25 |  |  | TO5 |
| 2N1711 | 75 | 30 | 7 | 1000 | 800 | 100-300 | 150 | 0.01 | $1 \cdot 5$ | 150/15 | 70 | 25 |  |  | TO5 |
| 2N2270 | 60 | 45 | 7 | 1000 | 1000 | 50-200 | 150 | 0.05 | 0.9 | 150/15 | 60 |  |  |  | TO5 |
| 2N3053t | 60 | 40 | 5 | 700 | 1000 | 50-150 | 150 | 0.25 | 1.4 | 150/15 | 100 |  |  |  | TO5 |
| 2N3704t | 50 | 30 | 5 | 800 | 300 | 90-330 | 50 |  | 0.6 | 100/5 | 100 | 12 |  |  | TO92a |
| 2N3705 | 50 | 30 | 5 | 800 | 300 | 45-165 | 50 |  | 0.8 | 100/5 | 100 | 12 |  |  | T092a |
| 2N3706 | 40 | 20 | 5 | 800 | 300 | 30-660 | 50 |  | 1.6 | 100/5 | 100 | 12 |  |  | T092a |
| 2N3794 | 40 | 20 | 5 | 300 | 250 | 100 min | 100 | 0.5 | 0.4 | 10/1 | 100 | 10 |  |  | U29 |
| 40361 † | 70 250 | 70 | 4 | 700 | 1000 | 70-350 | 50 | 0.25 | 1.4 | 150/15 | 100 |  |  |  | TO5 |
| 40412 | 250 |  |  | 1000 | 1000 | 40 min | 30 |  |  |  | 10 |  |  |  | TO5 |

Medium Current Audio, Small Output Types, PNP

| Type | Vebo max | $V_{\max }$ | Veb max | $I_{c}$ mA max | $\boldsymbol{P}_{\text {tot }}$ max | hFE <br> ${ }^{*} h_{f}$ | $I_{e}$ mA | $\\|_{\text {cbo }}$ $\mu$ | Vee (sat) (sat) | $\mathbf{l}_{c} / \mathrm{l}_{\mathrm{b}}$ mA | $\underset{\mathrm{MT}}{\mathrm{~Hz}}$ | $\begin{gathered} \mathbf{C}_{\text {ob }} \\ \mathbf{p F} \end{gathered}$ | $\begin{aligned} & N F \\ & d B \\ & \max \end{aligned}$ | $\begin{aligned} & \text { PG } \quad \text { MHz } \\ & \text { dB } \\ & \text { (a) } \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BC126 | 35 | 30 | 5 | 600 | 300 | 30-120 | 150 | 0.05 | 0.5 | 150/15 | $200 t$ | 5 |  |  | R097 |
| BC160-10 | 40 | 40 | 5 | 1000 | 750 | 63 min | 100 | 0.1 |  |  | $100 t$ | 20 |  |  | TO5 |
| BC303 | 90 | 65 | 7 | 1000 | 1000 | 40 min | 150 | 0.02 | 0.65 | 150/15 | 60 |  |  |  | TO5 |
| BFX29t | 60 | 60 | 5 | 600 | 500 | 50-200 | 10 | 0.075 | 0.4 | 150/15 | 100 | 12 |  |  | TO5 |
| BFX87 | 50 | 50 | 4 | 600. | 600 | 40 min | 10 | 0.05 | 0.4 | 150/15 | 100 | 12 |  |  | TO5 |
| BFX88 $\dagger$ | 40 | 40 | 4 | 600 | 600 | 40 min | 10 | 0.05 | 0.4 | 150/15 | 100 | 12 |  |  | TO5 |
| 2N1131 | 50 | 35 | 5 | 600 | 600 | 20-45 | 150 | 1 | $1 \cdot 5$ | 150/15 | 50 | 45 |  |  | TO5 |
| 2N1132 | 60 | 40 | 5 | 600 | 600 | 30-90 | 150 | 1 | $1 \cdot 5$ | 150/15 | 60 | 45 |  |  | TO5 |
| 2N3702t | 40 | 25 | 5 | 200 | 200 | 60-330 | 50 | $0 \cdot 1$ | 0.25 | 50/5 | 100 | 12 |  |  | TO92a |
| 2N3703 | 50 | 30 | 5 | 200 | 200 | 30-150 | 50 | 0.1 | $0 \cdot 25$ | 50/5 | 100 | 12 |  |  | TO92a |
| 2N4036 | 90 | 65 | 7 | 1000 | 1000 | 40-140 | 150 | $0 \cdot 1$ | 0.65 | 150/15 | 60 |  |  |  | TO5 |
| 2N4291 | 40 | 30 | 6 | 500 | 250 | 100-300 | 100 | 0.5 | 0.4 | 100/10 | 1504 | 10 |  |  |  |
| 40362† | 70 |  | 4 | 700 | 1000 | 35-200 | 50 |  | $1 \cdot 4$ | 150/15 | 100t |  |  |  | TO5 |

High Current Audio, NPN

| Type | $V_{\text {ebo }}$ max | $\mathbf{V}_{c}$ $\max$ | $\mathbf{V}_{e b}$ $\max$ | $\mathrm{I}_{\mathrm{max}}$ | $\begin{aligned} & P_{\text {tot }} \\ & \max \end{aligned}$ | $h_{\text {FE }}$ <br> ${ }^{\prime} h_{f}$ 。 | $\begin{gathered} \mathbf{l}_{\mathbf{c}} \\ \mathbf{m A A}^{(0)} \end{gathered}$ | Icbo $\mu \mathrm{A}$ | $\begin{aligned} & \mathbf{V}_{c \oplus} \\ & \text { (sat) } \end{aligned}$ | $I_{c} / I_{b}$ mA | $\mathrm{MT}_{\mathrm{Mz}}$ | $\begin{gathered} \text { Cob } \\ \text { PF } \end{gathered}$ | NF dB max | $\begin{aligned} & \text { PG MHz } \\ & \text { dB } \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BD121 | 60 | 35 | 6 | 5A | 303 | 30-100 | $1 \cdot 5 A$ | 50 | 0.65 | 1A/100 | $95 t$ |  |  |  | TO3 |
| BD123 | 90 | 60 | 6 | 5A | 303 | 30-100 | 1.5A | 50 | 0.65 | 1A/100 | 85t |  |  |  | TO3 |
| BD124 | 70 | 45 | 6 | 4A | 133 | 30-150 | 500 | 2 | 0.25 | 500/50 | 60 |  |  |  | MD17c |
| BD130 $\dagger$ | 100 | 60 | 7 | 15A | 666 | 20-70 | 4A | 5000 | $1 \cdot 1$ | 4A/400 | 1-1t |  |  |  | TO3 |
| BD131 $\dagger$ | 70 | 45 | 6 | 4A | 167 | 30-75 | 500 |  |  |  | 60 |  |  |  | $\times 58$ |
| BD135 $\dagger$ | 45 | 45 | 5 | 1.5 A | 100 | 40-100 | 150 | 0.1 | 0.6 | 500/50 | 50 |  |  |  | $\times 58$ |
| BD441 | 140 | 120 | 7 | 8A | 666 | 20-70 | 2 A |  | 1 | 2A/200 |  |  |  |  | TO3 |
| BDY20 | 100 | 60 | 5 | 15A |  | 20-100 | 4A |  | $1 \cdot 1$ | 4A/400 | 1 |  |  |  | TO3 |
| MJ481† | 60 | 60 | 5 | 4A | 500 | 30-200 | 1 A | 1000 | 0.4 | 1A/100 | 4 |  |  |  | TO3 |
| MJE521 | 40 | 40 | 4 | 3A | 320 | 40 min | 1A |  |  |  |  |  |  |  | $\times 58$ |
| MJE3055 | 70 | 60 | 5 | 10A | 718 | 20-70 | 4 A | 1000 | $1 \cdot 1$ | 4A/400 | 2 |  |  |  | $\times 58 \mathrm{c}$ |
| TIP31A | 60 | 60 | 5 | 3A | 320 | 20-100 | 1 A | 500 |  |  | 3 |  |  |  | X75b |
| 2N3054 | 90 | 55 | 7 | 4A | 142 | 25-250 | 500 | 1000 | 1 | 500/50 | $0 \cdot 75$ |  |  |  | TO66 |
| 2N3055 $\dagger$ | 100 | 60 | 7 | 15A | 660 | 20-70 | 4A | 5000 | $1 \cdot 1$ | 4A/400 | 0.4 |  |  |  | TO3 |
| 2N4915 | 80 | 80 | 5 | 5A | 500 | 25-100 | 2.5A | 100 | 1 | $2 \cdot 5 \mathrm{~A} / 250$ | 4 |  |  |  | TO3 |
| 2N5192 | 80 | 80 | 5 | 4A | 320 | 20-80 | $1 \cdot 5 \mathrm{~A}$ | 100 | $0 \cdot 6$ | 1-5A/150 | 2 |  |  |  | $\times 58$ |
| 40250 | 50 | 40 | 5 | 4A | 166 | 20-100 | 1.5A | 1000 | $1 \cdot 5$ | $1 \cdot 5 \mathrm{~A} / 150$ | 1-2t |  |  |  | TO66 |
| 40251 | 50 | 40 | 7 | 15A | 666 | 15-60 | 8 A | 5000 | 1.5 | 8A/800 | 0.5t |  |  |  | TO3 |

## High Current Audio, PNP

| Type | Vcbo max | $\begin{aligned} & \mathbf{V}_{c e} \\ & \max \end{aligned}$ | $V_{\text {eb }}$ max | $\underset{\max }{\mathbf{I}_{\mathbf{c}}}$ | $P_{\text {tot }}$ $\max$ |  | $I_{c}$ mA | Ifbo $\mu \mathrm{A}$ | $V_{c e}$ $(s a t)$ | $\mathrm{I}_{\mathrm{c}} / \mathrm{I}_{\mathrm{b}}$ mA | $\stackrel{\text { IT }}{\text { MHz }}$ | $\begin{gathered} C_{a b} \\ \mathbf{p F} \end{gathered}$ | NF dB max | $\begin{gathered} \text { PG MHz } \\ \text { dB } \\ \text { @ } \end{gathered}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BD132 $\dagger$ | 45 | 45 | 4 | 2 A | 167 | 35-75 | 500 |  |  |  | 60 |  |  |  | $\times 58$ |
| BD136 $\dagger$ | 45 | 45 | 5 | $1 \cdot 5 \mathrm{~A}$ | 100 | 40-100 | 150 | 0.1 | 0.6 | 500/50 | 50 |  |  |  | X58 |
| MJ491 $\dagger$ | 60 | 60 | 5 | 4A | 500 | 30-200 | 1 A | 1000 | 0.4 | 1A/100 | 4 |  |  |  | TO3 |
| MJE371 | 40 | 40 | 4 | 3A | 320 | 40 min | 1 A |  |  |  |  |  |  |  | X58 |
| MJE2955 | 70 | 60 | 5 | 10A | 718 | 20-70 | 4 A | 1000 | $1 \cdot 1$ | 4A/400 | 2 |  |  |  | $\times 58 \mathrm{c}$ |
| T1P32A | 60 | 60 | 5 | 3A | 320 | 20-100 | 1 A | 500 |  |  | 3 |  |  |  | $\times 75 \mathrm{~b}$ |
| 2N4906 | 80 | 80 | 5 | 5A | 500 | 25-100 | 2.5A | 100 | 1 | 2.5A/250 | 4 |  |  |  | TO3 $\times 58$ |
| 2N5195 | 80 | 80 | 5 | 4A | 320 | 20-80 | 1.5 A | 100 | $1 \cdot 4$ | 1-5A/150 | 2 |  |  |  | X58 |

See notes on power dissipation.

## Medium Current Switching, NPN

| Type | $\mathbf{V}_{\text {cbo }}$ max | $\begin{aligned} & \mathbf{V}_{\text {ce }} \\ & \max \end{aligned}$ | $V_{e b}$ $\max$ | $\underset{\max }{\boldsymbol{I}_{c}}$ | Ptot max | $h_{\text {FE }}$ "hfe $^{\prime}$ | $J_{c}$ mA | Ifebo $\mu \mathrm{A}$ | $\begin{aligned} & V_{\text {ee }} \\ & (s a t) \end{aligned}$ | $1 \mathrm{c} / \mathrm{f}_{\mathrm{b}}$ mA | $\stackrel{f T}{\mathrm{H}_{\mathrm{Hz}}}$ | Cob pF | NF $d B$ <br> max | $\begin{aligned} & \text { PG MHz } \\ & \mathbf{d B} \end{aligned}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BSX20 $\dagger$ | 40 | 15 | $4 \cdot 5$ | 500 | 360 | 40-120 | 10 | 0.4 | 0.6 | 100/10 | 600 |  |  |  | TO18 |
| C762 $\dagger$ | 40 | 25 | 5 | 150 | 400 | 30-110 | 50 | 0.2 | 0.6 | 150/15 | 350 | 5 |  |  | T018 |
| MPS6531 | 60 | 40 | 5 | 600 | 310 | 90-270 | 100 | 0.05 | 0.3 | 100/10 | 390 t | $3 \cdot 5$ |  |  | TO92b |
| P346A | 25 | 12 | 3 |  | 300 | 20 min | 10 | 0.05 | 1 | 100/1 | 400 | 4 |  |  | TO18 |
| 2N706 | 25 | 20 | 3 | 200 | 300 | 20 min | 10 | 0.05 | 0.6 | 10/1 | 200 | 6 |  |  | T018 |
| 2N706A | 25 | 15 | 5 |  | 300 | 20-40 | 10 | 10 |  |  | 200 | $3 \cdot 5$ |  |  | T018 |
| 2N2218 | 60 | 30 | 5 | 800 | 800 | 40-120 | 150 | 0.01 | $1 \cdot 6$ | 500/50 | 250 | 8 |  |  | TO5 |
| 2N2218A | 75 | 40 | 6 | 800 | 800 | 40-120 | 150 | 0.01 | 1 | 500/50 | 250 | 8 |  |  | TO5 |
| 2N2219 | 60 | 30 | 5 | 800 | 800 | 100-300 | 150 | 0.01 | $1 \cdot 6$ | 500/50 | 250 | 8 |  |  | TO5 |
| 2N2219A | 75 | 40 | 6 | 800 | 800 | 100-300 | 150 | 0.01 | 1 | 500/50 | 250 | 8 |  |  | T05 |

## Medium Current Switching, PNP

| Type | $V_{\text {cbo }}$ max | $\begin{aligned} & \mathrm{V}_{\mathrm{ce}} \\ & \max \end{aligned}$ | $\begin{aligned} & V_{e b} \\ & \max \end{aligned}$ | Ic $m A$ $\max$ | $\begin{aligned} & P_{\text {tot }} \\ & \text { max } \end{aligned}$ | $h_{\text {FE }}$ <br> *hfe | $\begin{aligned} & \mathbf{I}_{\boldsymbol{c}} \\ & \mathrm{mA} \end{aligned}$ | $I_{c b o}$ $\mu \mathbf{A}$ | $\begin{aligned} & \mathbf{V}_{\text {ce }} \\ & \text { (sat) } \end{aligned}$ | $I_{c} / l_{b}$ mA | ${ }_{\mathrm{M}}^{\mathrm{H}} \mathrm{~Hz}$ | $\begin{gathered} \text { Cob }_{\text {pr }} \\ \text { pF } \end{gathered}$ | NF dB max | $\mathbf{P B}_{\text {(a) }}$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MPS6534 | 40 | 40 | 4 | 600 | 310 | 90-270 | 100 | $0 \cdot 15$ | 0.3 | 100/10 | 250 | 6 |  |  | TO92b |
| V763 $\dagger$ | 25 | 25 | 5 |  | 400 | 30 min | 50 | 0.05 | 1 | 150/15 | 300 | 6 |  |  | T018 |
| 2N2904 | 60 | 40 | 5 | 600 | 600 | 40-120 | 150 | 0.02 | 0.4 | 150/15 | 200 | 8 |  |  | TO5 |
| 2N2904A | 60 | 60 | 5 | 600 | 600 | 40-120 | 150 | 0.01 | 0.4 | 150/15 | 200 | 8 |  |  | TO5 |
| 2N2905 | 60 | 40 | 5 | 600 | 600 | 100-300 | 150 | 0.02 | 0.4 | 150/15 | 200 | 8 |  |  | TO5 |
| 2N2905A | 60 | 60 | 5 | 600 | 600 | 100-300 | 150 | 0.01 | 0.4 | 150/15 | 200 | 8 |  |  | TO5 |


| $\longdiv { M D 1 7 c }$ |  | RO9 | R097 | RO97a |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { T05 } \\ & \overbrace{0}^{e} \mathrm{t} \\ & \hline \end{aligned}$ |  |  | $1066$ |  |  |  |
|  | U29 | $\times 9 a$ |  | $\begin{array}{\|c} \hline 0 \\ 058 a, c \\ \text { ecb } \\ \text { mear } \\ \text { mat to } \\ \text { coilector } \end{array}$ |  | $\left\{\begin{array}{c} \times 75 \\ \underbrace{}_{\substack{\text { IIf } \\ \text { bce } \\ \text { mounting lug } \\ \text { mo collector }}} \\ \text { to } \end{array}\right.$ |

I$T$ is an unwritten law of electronics that whenever a group of engineers are gathered together then sooner or later a HiFi amplifier will be designed. Usually they, the amplifiers, that is, are assembled in such a way that even if they work well at the time they are quite impossible to reproduce.

The Texan is mainly a 'value-for-money' design which the average home constructor, armed with little more than a soldering iron, should be able to build successfully in a time which will allow him to remain on speaking terms with his wife.

The original design was produced for a housepublication* and was intended to promote the use of integrated circuit operational amplifiers for audio use. Also, since I have an aversion to printed-circuit boards bristling with capacitors-particularly electro-lytics-it was designed to reduce these to an absolute minimum.

The present circuit is updated to use the latest devices and it includes extra facilities such as scratch and rumble filters.

The printed circuit board has also been completely changed so that most of the controls are mounted directly and very little spaghetti wiring is required.

## Why use operational amplifiers?

By definition the main function of a preamplifier is to provide gain and to do it in a way which is well controlled with respect to frequency response and stability.


Fig. 1: (a) Triangle represents a device which has infinite voltage gain.
(b) With two resistors added.
(c) Modlfied circuit-see text for detalled explanatlon.

If someone made a transistor with a potential voltage gain 200,000 , an $\mathrm{f}_{\mathrm{T}}$ of 2 MHz , a maximum phase shift of $90^{\circ}$, differential inputs with high common mode rejection and so and so forth there would be little point in using an operational amplifier, but to the best of my knowledge they do not. However, operational amplifiers are now so easy to use that they can be treated simply as a transistor with eight legs.
More explicity the operational amplifier has the following features:
(1) A very high mid-band gain which allows accurate equalization while retaining sufficient loop gain to reduce distortion to very low levels.
(2) Inherently high input impedance and low output impedance which make feedback networks simple to design and minimise the effect of loading.
(3) Large potential output swing giving good overload ability.
(4) Balanced design with d.c. coupling and very low offset voltages. This greatly reduces the number of electrolytic capacitors required in the amplifier and allows the output to operate without the need for manual setting up of the d.c. output conditions.
(5) High supply ripple voltage rejection which means that the operational amplifiers will work on poorly regulated supplies and reduces the likelihood of low frequency instability, one of the most common faults in many amplifier designs.
(6) Good common-mode rejection ratio of about 90 dB . This describes the ability of the operational amplifier to see only the differential input and ignore voltages which are common to both input terminals so that spurious voltages appearing along a length of printed track, for example, can be almost totally rejected provided both inputs are referred to the same point on the track.

## Operational amplifier theory

If you are familiar with operational amplifier theory skip the following. If not, read on.
Operational amplifiers are a very close approach, for most purposes, to the mythical 'black box' for which we make the following assumptions.

The triangle shown in Fig. 1a is a device which has infinite voltage gain (A) for differential voltages appearing across the $+_{\text {in }}$ and - in terminals, for which no current flows into the inputs and whose output impedance is zero.


ERED
richard mann*


Fig. 2: One complete channel of the "Texan" stereo amplifier.

* Texas Instruments Limited.

Therefore, if we add two resistors as in Fig. 1b, since the + input is earthed and $A=\infty$ there is no difference in voltage between the terminals and the - input stays at ground potential becoming a 'virtual earth'. The current through R1, must therefore be $\mathrm{V}_{\mathrm{in}} / \mathrm{R} 1$.

As no current flows into the 'box' all input current must flow through R2 producing a voltage across it of $\frac{V_{i n}}{R I} R 2$ and since the current flows from the virtual earth point, $V_{\text {out }}$ must be

$$
\left(0-\frac{V_{i n}}{R 1} R 2\right)
$$

Therefore, the closed loop gain of the circuit is

$$
\frac{V_{\text {out }}}{V_{\text {in }}}=-\frac{R 2}{R 1}
$$

If we modify the circuit as shown in Fig. 1c, we have a slightly different situation. The voltage at both input terminals becomes $+V_{i n}$ so the feedback current is still $\frac{V_{\text {in }}}{R 1}$ but it now flows in the opposite direction so that $\mathrm{V}_{\text {out }}=\frac{\mathrm{V}_{\mathrm{in}}}{\mathrm{R} 1}(\mathrm{R} 1+\mathrm{R} 2)$ and the closed loop gain $=\frac{V_{\text {out }}}{V_{\text {in }}}\left(\frac{R 1+R 2}{R 1}\right)$

These two configurations are the classic 'inverting' and 'non inverting' amplifiers. The main differences, apart from polarity, are the input impedances, which for the former is R1 and for the latter is $\infty$, and the fact that the non-inverting amplifier cannot have a gain of less than unity.

## The SN72748P

Most integrated circuit manufacturers make at least one version of the popular '741' operational amplifier design. This ingenious device was designed to improve on the equally popular ' 709 ' which in turn was a vast improvement on earlier integrated circuits which were essentially monolithic versions of discrete component designs. Although the 741 now has competitors from the new range of 'superbeta' amplifiers it is likely to remain an industry standard for a long time mainly because it offers very good all round performance at a very low price.

The SN72748P and SN72741P are fairly typical scions of the 741 family. The difference between them is merely that the SN72741P has internal frequency compensation which allows $100 \%$ negative feedback to be applied without the circuit becoming unstable. The SN72748P requires one small capacitor (about 10 pF ) to be added externally and gives a higher gain-bandwidth than the 741. Physically, they are identical in 8 pin dual-in-line plastic packages.

Typical parameters for the SN72748P when operated at $\pm 15 \mathrm{~V}$ supplies are:
Input resistance: $2 \mathrm{M} \Omega$ (higher with feedback. Output resistance: $75 \Omega$ (much lower with feedback). Gain: 200,000.
Input offset voltage ( $\mathrm{V}_{\mathrm{IO}}$ ): 1mV (this represents the matching error of the input transistors).
Maximum output current: 15 mA r.m.s.
Maximum output voltage: 9V r.m.s.
Input bias current ( $\mathrm{I}_{\text {IB }}$ ): $0 \cdot 080 \mu \mathrm{~A}$ (this is the current

## $\star$ Specification

Sensitivity to give 20 watts into $8 \Omega$
Radio 30 mV
Magnetic pickup 2.5 mV
Auxiliary (see text)

## Input impedance $47 \mathrm{k} \Omega 1 \mathrm{kHz}$

Tape output (low level output) 130 mV via $4 \cdot 7 \mathrm{k} \Omega$ unaffected by tone or volume controls

## Tone controls

Treble $+10-12 \mathrm{~dB}$ at 15 kHz
Bass $\pm 16 \mathrm{~dB}$ at 30 Hz
Balance $\pm 8 \mathrm{~dB}$ one channel relative to other

## Filters

Rumble-critically damped 2nd order corner frequency $50 \mathrm{~Hz}, 12 \mathrm{~dB}$ per octave roll-off
Scratch-critically damped 2nd order corner frequency $5 \cdot 5 \mathrm{kHz}, 12 \mathrm{~dB}$ per octave roll:off
Interchannel crosstalk
-51 dB at 1 kHz
-48 dB at 10 kHz
Unweighted signal-to-noise of complete amplifier with full amplifier bandwidth
-60 dB magnetic pickup
-72 dB radio
Dynamic range (equalisation and tone control stages)
-38 dB before clipping (above nominal input level)
Power Output (both channels)
$20+20$ watts into $8 \Omega$ intermittent sine wave
$16+16$ watts into $8 \Omega$ continuous sine wave
$15+15$ watts into $15 \Omega$ continuous sine wave

## Harmonic distortion

15 watts into $15 \Omega \quad 0.05 \%$ at 1 kHz
20 watts into $8 \Omega \quad 0.09 \%$ at 1 kHz
Low level distortion
$0.16 \%$ at 1 kHz 50 mW into $15 \Omega$
$0.07 \%$ at 1 kHz 50 mW into $8 \Omega$
Intermodulation distortion-wave analyser plots will appear later in this series
Frequency response ( $16+16$ watts into $8 \Omega$ )
-1 dB 7 Hz to 22 kHz
-3 dB less than 5 Hz to 35 kHz
Stabillty-will drive electrostatic loudspeakers
Output impedance less than 1 milli ohm
Dimensions (less teak sleeve)
width 363 mm 14.3 in
height 48 mm 1.9 in
depth 149 mm 5.9 in
The following equipment was used to obtain the above measurements on the TEXAN amplifier:

Low Distortion Oscillator--Radford. Wave AnalyserHewlett Packard 3590A/3593A. True R.M.S. VoltmeterHewlett Packard. Function Generator-Hewlett Packard 3300A/3305A. A.C. Digital Voltmeter-Pacific Measurements 1010. Oscilloscope_Hewlett Packard 181 A. Test Oscillators (two)-Hewlett Packard 652A (intermodulation tests). X Y Recorder-Hewlett Packard 136A.
necessary to turn on the input transistors).
Supply ripple rejection: 20,000:1
More comprehensive details of the devices, including circuitry, can be obtained from manufacturers' data sheets.

## Description of circuit

The complete amplifier circuit is given in Fig. 2, but components are referred to below for only one channel. In the components list, items for the other channel are +100 (i.e. R1, R101).
The input stage amplifies the various input signals to approximately 50 mV , provides whatever equalization is required and gives the loop gain necessary for an active rumble filter. The feedback may look rather complex, but it breaks down into three parts as shown below.
(1) At d.c. the output is returned to the inverting input via $\mathbf{R}$ (Fig. 3a) which comprises R4 in parallel with R5, R6 or R9.
C3 blocks the path to ground so substituting in the equation for Fig. 1 C we have a d.c. gain of $\frac{\mathrm{R}+\infty}{\infty}$ (i.e. unity).

Since R 2 refers the + ve input to ground, the -ve input will sit at a potential of $\mathrm{V}_{\text {IO }}$ (approx $\pm 1 \mathrm{mV}$ ) and the output will also take this d.c. level. This offset is so low that there is no need for a coupling capacitor into the next stage.
(2) At mid band a.c. the gain of the input stage becomes $\frac{\mathrm{Z}+\mathrm{R} 3}{\mathrm{R} 3}$ where Z (Fig. 3b) is the impedance of the selected feedback network.
R4 has no effect on $Z$ since the current through R4 is shunted to ground by C3. The feedback networks shown are designed to handle radio, magnetic pickup and auxiliary inputs and are discussed in more detail later in this series of articles.

The input impedance of the stage is inherently very high ( $>2 \mathrm{M} \Omega$ ) but is shunted by the $47 \mathrm{k} \Omega$ resistor R 2 which provides the damping necessary for magnetic pickups and supplies the minute input bias current for the operational amplifier,
(3) A classic form of second order, high-pass active filter is shown in Fig. 3c.

The amplifier is ideally a unity gain buffer. Being a second order filter the low frequency response tends to roll off at $40 \mathrm{~dB} /$ decade and the transitional region is a function of the damping factor () of the

(a)

(c)

(b)

Fig. 3: (a) D.C. conditions of the input stage.
(b) At mid-band a.c.
(c) Classic form of 2nd order high-pass act/ve filter.
(see text for detalled explanation).

(a)

Fig. 4 : (a) Theoretical response of 2nd order high pass filter for different values of $\zeta$
filter. The effect of $\varsigma$ is shown in Fig. $4 a$ and has optimum cut-off without peaking when $\zeta=\frac{1}{\sqrt{2}}$. The damping factor is evaluated from $\zeta=\sqrt{\frac{\mathrm{R} 1}{\mathrm{R} 2}}$ and cutoff frequency ( $\mathrm{f}_{0}$ ) from $\mathrm{f}_{0}=\frac{1}{2 \pi \sqrt{\mathrm{R} 1 \mathrm{R} 2 \mathrm{C} 1 \mathrm{C} 2}}$

Since the operational amplifier has to supply a certain amount of gain in the equalization stage its output obviously does not provide a unity gain buffer. However, the inverting input terminal can be used for this purpose since its voltage exactly follows that of the non-inverting terminal.

The characteristic of the filter is modified slightly because any current flowing down R1, is not shunted to ground via R3 and C3, as one might think, but has to flow through the feedback impedance $Z$. Nevertheless the response is very close to a standard filter response as shown in Fig. 4b.

Placing the filter right at the front of the amplifier ensures that large inputs at sub-audio frequencies will not cause any intermodulation in ICl.

The push button switch $S 1$ removes the filter when the button is out and a direct connection is made to the selected input.

This is permissable since the d.c. bias current to the operational amplifier is in the order of nano amps and causes no problems in the average pickup or tape head.

Facilities for coupling the equalised input signal to an external tape recorder are provided by taking the output of 1 Cl via a $4 \cdot 7 \mathrm{k} \Omega$ resistor R10. This


Fig. 4 : (b) Rumble filter characteristic (measured).
series resistance allows the channels to be shorted externally if only a mono signal is required. The very low output impedance of IC1 $(<1 \Omega)$ means that no significant crosstalk is introduced in these circumstances.

## The tone control stage

The tone control is a standard Baxandall circuit using an SN72741P to provide the loop gain. A fully compensated amplifier is necessary here since there is $100 \%$ feedback around the circuit in the cut position.
The time constants of the feedback network are chosen so that there is negligible interaction between the bass and treble controls. The control range is shown in Fig. 5.
The maximum levels of bass boost and cut are determined by the end stop resistors R13 and R16 respectively. Similarly the range of the treble control is limited by R12 and R15 although in this case the bass network does have a modifying influence which adds an extra limiting factor.

The transient response of the tone control with maximum treble boost is shown in the oscillogram, Fig. 6. It is well damped and shows no tendency to ring.

Two other functions are included in the tone control stage, the mono-stereo switch and the balance control.
Fnding the optimum position for the mono switch was rather a problem since it should be before
the balance control if possible. The series resistor Rll was added so that the two channels could be shorted together by $S 2$ without overloading the preamplifiers (ICl IC101). However, the resistor is not high enough to have a significant effect on the tone control range with the value given.
The balance control is an active circuit giving a control range of $\pm 12 \mathrm{~dB}$ for one channel relative to the other. This is a personal preference as I do not see much point in having $100 \%$ control range and it usually involves a waste of gain. However, anyone feeling strongly about it could cut out the relevant components and insert a dual gang pot between Cl 3 and the volume control. The $10 \mu \mathrm{~F}$ capacitor C 13 , incidentally, is the only d.c. block in the forward signal path which accounts for the good low frequency response and phase shift of the amplifier. A tantalum 'bead' capacitor is used in this position partly for its small size and partly because they will withstand a reverse polarizing voltage of 0.5 V . Because the output voltage of IC2 depends mainly on its input offset it could be of either polarity but its magnitude will not exceed 200 mV for any setting of the bass control (which determines the d.c. gain of the stage). A minor point here is that if the bass control is moved rapidly, a transient level shift may be heard from the loudspeaker. This effect can be removed by inserting a blocking capacitor of about $1 \mu \mathrm{~F}$ between R13 and the junction of Cl0 and VR2, thereby fixing the d.c. gain of the stage at unity.

## The scratch filter

The scratch filter is again based on a classic second order configuration as shown in Fig. 7. This lowpass circuit is the dual of the high-pass circuit used for the rumble filter (Fig. 3c). The damping factor is $\sqrt{\overline{\mathrm{Cl}}}$ and the cut-off frequency $\frac{1}{2 \pi \sqrt{\mathrm{C} 1 \mathrm{C} 2 \mathrm{R} 1 \mathrm{R} 2}}$

## TRANSISTORS

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| AAF30 | 10 D | BD115 | 75p |
| :---: | :---: | :---: | :---: | :---: |
| AAY42 | OC16 |  |  | | AAY42 | 15 D | $\mathrm{BDIL3}$ | 85 p | OC20 |
| :--- | :--- | :--- | :--- | :--- | | $A A Z 13$ | 109 | BD124 | 80 p | Oc22 |
| :--- | :--- | :--- | :--- | :--- |

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 | AC128 | 20 p | BD |
| :--- | :--- | :--- |
| AC176 | 20 p | BD |



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D148
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D182$A D 162$
AF114
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$\begin{array}{ll}18118 \\ \mathrm{~F} 124 & 26 p \\ \mathrm{Fi25} & 20 p \\ \mathrm{FI} 128 & 200\end{array}$
F139
F180
F181
F185
F186

| 186 | 40 | B8T95A | $12 p$ | 8 |
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| 8995 | 120 | $T$ |  |  |

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| 110 | $26 p$ | BYZ13 | $80 p$ |
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| 115 | 70 | BZY78en |  |

AX19 59 GZY78A100

| AX16 |
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Triple 3 －input NAND gates
4－input Schmitt triggers
7420 Dual 4 －input NAND gates
7430 Sligle 8－input NAND gates
7440 Dual 4－input NAND butfer gates
$7441 \quad$ BCD－Decimal decoder／Nlie driver
748 Excess 3－Decimal decoder TTL outputs
$\begin{array}{ll}7447 & \text { BCD－Decimal } 7 \text { aeg．decoder／indicator driver } \\ 7448 & \text { BCD－Decimal } 7 \text { eeg．decoder／driver TTL O／P }\end{array}$
7450 Expand dual 2－input AND－OR－INVERT gates
7451 Dual 2－wide 2－input AND－OR－INVERT
7451 Dual 2－wlde 2－input AND－OR－INVERT gates
7458 Quad 2－input expand AND－OR－INVERT gates
7458 Quad 2－input expand AND－OR－INVERT g
7454
7 －wide 2－input AND－OR－INVERT gates
7460 Dusi 4－input expanders
7470 single J－K flip－flop（gated Inputa）
7472 Single J－K ilp fiop（gated inputs）
Dual J－K flip flop
Dual D fit fiop
Qumiruple blatable latci
7478 Dual J－K flip－flops with Preset and Clear
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7489
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7482
2－blt binary Full Adder
$\begin{array}{ll}7388 & \text { 2－blt binary Full Adder } \\ 7488 & \text { 4－bit binary Full Adder }\end{array}$
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7488 Quadruple 2－input Exclueive OR gate．
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7491 g－blt abift reglater
7498 Divide twelve counter
7482 Divide twelve count
4－blt blary counter
7494 Dual entry 4 －bit shift reqiater
7406 4－blt up－dowd abift register
2490 5－bit parallel／serial in／out abift register

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74181 Hextuple Bet－Reset latches
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 75151 16－bit data relector／multiplexer
75151 8－bit data selector／muitiplexer
74154 16－bit decoder／demultiplexer
74155 Dual 2 －line to 4 －line decoder／denultiplexer
74190 Sync decade up－down connter，1－itne mode
 74198 sync 4－bit up－down counter， 2 －line mode
74100 Asynchronous presettable decede counter
74197 Aaynchronous preaettable 4 －blt blnary counter

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| 201 | 189 | 149 | 14p |
| 20 p | 189 | 16p | 140 |
| 20 p | 18p | 10p | 145 |
| 20 p | 18p | 16 | $14 \%$ |
| 20p | 189 | 16 | 149 |
| 20 p | 18p | 180 | 140 |
| 20p | $18 p$ | 16 p | 140 |
| $30 p$ | 278 | 2 p | 2909 |
| $20 p$ | 189 | 160 | 14 |
| 20p | $18 p$ | 16 p | 149 |
| 20 p | 18p | 16p | 149 |
| 760 | 72 | 700 | $00 p$ |
| 750 | 789 | 700 | 60p |
| 11．00 | 95 p | 90 p | 801 |
| 21.75 | 81.60 | 81.46 | \＄1．80 |
| 81.75 | 51.60 | 11.45 | 11.30 |
| 20 p | 18p | 10p | 149 |
| 20 p | 18p | 160 | 148 |
| 20 p | 189 | 160 | 149 |
| 20 p | 18p | 16p | 14 |
| 200 | 18p | 169 | 14p |
| 300 | 27 | 85p | 28 p |
| $80 \%$ | 87 | 250 | 20p |
| 409 | 870 | 80 | 385 |
| 400 | 87 p | 850 | 88 |
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| 400 | 870 | 84 | $81 p$ |
| $80 p$ | 789 | 67 | 58 |
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| 879 | 80 | 70p | 650 |
| 21.00 | 909 | 85 | 80 p |
| 909 | $85 p$ | 80 | 75p |
| 15p | 410 | 889 | 859 |
| 750 | 70 | 65p | 800 |
| 41.00 | 859 | 905 | 800 |
| $75 p$ | 700 | 06p | 600 |
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| 80 | 780 | $70 p$ | 650 |
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| 81.00 |  | $80 p$ | 80 p |
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| \％1．00 | 859 | 90 p | 600 |
| 81.50 | 81． 20 | 81.80 | 81.10 |
| 88.85 | \％ 80 | 28．95 | 20－15 |
| 81．10 | 85 | 000 | 80 p |
| 41.85 | 81.27 | 21.80 | t1．15 |
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| 11－55 | E1．47 | 81.85 | 81.10 |
| 21．55 | 51.47 | 81.85 | 81.10 |
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| TAD1 10 | 21.50 |
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| 709 C （TOS） | 450 |
| 709C（D．I．L．） | 45p |
| $723 \mathrm{CTO5}$ | 81.00 |
| 741C（TO5） | $80 p$ |
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| MC1304P | 28.25 |
| 8L403D | 81.60 |
| 741CDIL） | 75p |
| 914 （TOS） | $40 p$ |
| 923 （TOB） | 400 |
| T08EIBA |  |
| 20 watt amp． | 24．47 |
| TOEFIBA |  |
| Pre amp | 81.50 |



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Smp
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| 8C45B | 200 | 10 amps |
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$\begin{array}{llll}\text { 8C45B } & 200 \quad 10 \mathrm{amps} & \mathbf{5 1} \cdot 15 \\ \text { BCA5D } & 400 \quad 10 \mathrm{ampa} & 81 \cdot 25\end{array}$
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R3 R103 $1 \mathrm{k} \Omega 5 \%$ iW
R4 R104 $100 \mathrm{k} \Omega$
R5 R105 1-2k 5\% ${ }^{\frac{1}{4} W}$
R6 R106 270ks 5\% $\mathbf{1}$ W
R7 R107 $22 \mathrm{k} \Omega 5 \%$ W
R8 R108 - $5 \%$, $\mathbf{~ W}$ R9 R109 1-2k $5 \%$ 交W $\}$
R10 R110 $4 \cdot 7 \mathrm{k} \Omega$
R11 R111 $1 \cdot 8 \mathrm{k} \Omega$
R12 R112 3.3k $\Omega$
R13 R113 10k $\Omega$
R14 R114 $33 \mathrm{k} \Omega$
R15 R115 3.3k $\Omega$
R16 R116 $10 \mathrm{k} \Omega$
R17 R117 $1 \mathrm{k} \Omega$
R18 R118 2-2k $\Omega$
R19 R119 $22 \mathrm{k} \Omega 5 \%$ \% $W$
R20 R120 $22 \mathrm{k} \Omega 5 \%$ W
R21 R121 $680 \Omega 5 \%$ WW
All resistors $10 \% \underset{\mathrm{t}}{ } \mathrm{W}$ high stability carbon film uniess otherwise specified.

Potentiometers:
VR1 VR101 $100 \mathrm{k} \Omega$ twin-gang carbon linear VR2 VR102 $100 \mathrm{k} \Omega$ twin-gang carbon linear VR3 $5 \mathrm{k} \Omega$ single-gang carbon linear VR4 VR104 $10 \mathrm{k} \Omega$ twin-gang carbon log All pots have p.c. terminations-AB Metals type D45

| Capacitors |  |  |
| :---: | :---: | :---: |
| C1 C101 | $0.1 \mu \mathrm{~F}$ |  |
| C2 C102 | $0.1 \mu \mathrm{~F}$ |  |
| C3 C103 | $100 \mu \mathrm{~F}$ | 3 V tantalum |
| C4 C104 | $0 \cdot 1 \mu \mathrm{~F}$ |  |
| C5 C105 | 10pF 10 | \% 30V polystyrene |
| C6 C106 | $0.1 \mu \mathrm{~F}$ |  |
| C7 C107 | 3900pF |  |
| C8 C108 | $0.01 \mu \mathrm{~F}$ |  |
| C9 C109 | not no | ally required (see text, next month) |
| C10 C110 | $0.05 \mu \mathrm{~F}$ |  |
| C11 C111 | 560pF | $5 \% 30 \mathrm{~V}$ polystyrene or ceramic |
| C12 C112 | $0.05 \mu \mathrm{~F}$ |  |
| C13 C113 | $10 \mu \mathrm{~F}$ | 16 V tantalum |
| C14 C114 | 1500 pF | $5 \% 30 \mathrm{~V}$ polystyrene |
| C15 C115 | 1000 pF | $5 \% 30 \mathrm{~V}$ polystyrene |
| C16 C116 | $100 \mu \mathrm{~F}$ | 3 V tantalum |
| C17 C117 | 10pF | 10\% 30V polystyrene |
| C18 C118 | 47pF | 10\% 30V polystyrene |
| C19 C119 | $0 \cdot 1 \mu \mathrm{~F}$ |  |
| C20 | $1000 \mu \mathrm{~F}$ | 25V electrolytic Daly |
| C21 | $1000 \mu \mathrm{~F}$ | 25 V electrolytic Daly |
| C22 | $3500 \mu \mathrm{~F}$ | 50 V electrolytic Daly |
| C23 | $3500 \mu \mathrm{~F}$ | 50 V electrolytic Daly |
| All capacitors $10 \% 30 \mathrm{~V}$ polyester or mylar, unless |  |  |
| otherwise | specified |  |

## Switches:

$\left.\begin{array}{ll}\begin{array}{l}\text { S1 } \\ \text { S2 }\end{array} & 4 \text { pole } 2 \text { way } \\ \text { Sole } 2 \text { way }\end{array}\right\}$ push-on-push-off 1 pole 3 way make-before-break rotary

## Integrated Circuits \& Semiconductors:

IC1 IC101
IC2 IC102
IC3 IC103
Tr1 Tr101
Tr2 Tr102
Tr3 Tr103
Tr 4 Tr104
Tr5 Tr105
D1
D2
D2 D4
ZD1
ZD2


Indicator:
N1 Neon mains indicator
Transformer:
T1 Mains transformer primary 240 V secondary 20-0-20V 1A. (special design) Gardners SL20

## Sockets:

SK1 5W DIN socket
SK2 5W DIN socket
SK3 5W DIN socket
SK4 speaker DIN socket
SK5 speaker DIN socket
J1 3 pole stereo jack socket
Fuses:

| uses: |  |  |
| :--- | :--- | :--- |
| F1 F101 | 2A | $1 \cdot 25 \mathrm{in}$. cartridge |
| F2 F102 | 2A | $1 \cdot 25 \mathrm{in}$. cartridge |
| F3 | 1A | 20 mm anti-surge |

## IC Sockets:

8 pin DIL. IC sockets (IC sockets must be usedit is a condition of IC guarantee from Henry's Radio Ltd.)

Miscellaneous:
2 off twin fuseholders (1-25In fuses); 1 off panel mounting 20 mm fuseholder.

All components, including a ready drilled fibre glass printed circuit board, drilled and punched metaiwork, finished.front panel, will be available to home constructors and the trade, from Henry's Radio Ltd. A teak sleeve will aiso be availablemore details next month.

As with the rumble filter, its response is modified slightly by the feedback impedance of the output stage which provides its loop gain. The response of the scratch filter is given in Fig. 8 and shows that the circuit is again critically damped and reaches a roll off of nearly $40 \mathrm{~dB} /$ decade( $12 \mathrm{~dB} /$ octave).

## The output stage

The configuration of the output stage is almost identical to that of the input stage with a few extra watts thrown in. The midband gain of the stage is $\frac{\mathrm{R} 22+\mathrm{R} 21}{\mathrm{R} 21}$ and the reactance of C 16 reduces this
to unity at d.c. so that the final offset of the power stage is only 3 or 4 mV . This feature is of importance first because it removes the need for a level setting adjustment and secondly because it allows all normal loudspeakers to be connected directly without the need for a bulky blocking capacitor. Even a Quad electrostatic loudspeaker can be driven satisfactorily in spite of the low primary resistance of its matching transformer.

The power stage is rather unusual in that it provides about 20 dB voltage gain unlike the normal Darlington or other $100 \%$ feedback configuration.
The gain is introduced by attenuating the feedback with R31 and R28 for positive excursions and


Fig. 6: Transient response of the tone control with maximum treble


R32 and R29 for negative excursions. Voltage gain is necessary because the output swing of the operational amplifier is limited partly by its d.c. swing capability but mainly by its slew rate-the maximum rate in $V / \mu S$ at which the output can change.
The transistors used in the power stage are low cost plastic encapsulated types. The drivers are a complementary pair of Texas Silect devices type BC182 and BC212. They have a continuous rating of 300 mW in free air at $25^{\circ} \mathrm{C}$ which is quite adequate for driving the output transistors providing ballast resistors R30 and R33 are added to reduce their collector dissipations.

## Quiescent current setting

This is always a knotty problem in class B or class AB amplifiers and is a process which frequently means instant death to the output transistors. This does not matter too much if you work for the firm which makes them but it is not so good if you have parted with hard earned cash. For this reason the circuit shows fixed resistors (R24 and R25) which set-up the potential across the $\mathrm{V}_{\mathrm{be}}$ multiplier ( $\operatorname{Tr} 1$ ). This potential, in conjunction with the d.c. feedback introduced by the emitter resistors R28 and R29, sets the current in the driver stage and hence, in the output transistors. If you are a purist you can sit down with a slide rule and data sheets and prove that this will give you a very wide variation of quiescent current. However, a more empirical
approach based on experience with about two dozen amplifier boards and transistors from different batches have indicated that the current usually sets up between 10 mA and 40 mA and anywhere in this range gives good crossover performance. This is partly due to the current, rather than voltage drive into the bases of the output transistors which have no shunt resistors to their emitters. Omitting these resistors slows down the power devices slightly but this is allowed for in the frequency compensation. The oscillograms in Figs 9 a and 9 b show the smooth crossover for an output power of 100 mW r.m.s. with triangular inputs at 1 kHz and 10 kHz . Should the purist remain unconvinced, however, there is provision for a $470 \Omega$ preset potentiometer on the printed circuit board. This is wired in place of R28 which should be omitted if the preset is used. There is also a link between the emitter of each TIP42A and the +25 V rail which can be broken to give a current monitor point. This quiescent current in the output stage can then be set to 20 mA .

## Frequency compensation

It is vital to have an amplifier correctly compensated if good transient response is expected from an amplifier. All too often designs appear to have marginal stability which causes the home constructor endless problems. In the Texan a dominant pole is placed in the open loop response by the 10 pF capacitor, Cl7. A further pole occurs naturally in the discrete stage which is cancelled by a zero introduced with the phase advance capacitor, C 18 , This gives nearly $90^{\circ}$ of phase margin to the amplifier which allows it to cope quite well with highly capacitive loads. Fig. 10 shows the transient response when driving 20 watts into an $8 \Omega$ resistor in parallel with $2 \mu \mathrm{~F}$-said to be equivalent to an electrostatic loudspeaker load. There is a small overshoot under these conditions since transistor parameters have an inconvenient habit of changing with collector current. However, the degree of overshoot is not enough to cause trouble.
A similar oscillogram, Fig. 11, shows the output waveform when driving a Quad electrostatic loudspeaker, The power level in this case was only 1 W as higher levels proved rather mpopular with my colleagues!

In addition to the compensation components there is a dummy h.f. load (C19 and R36) which is not necessary when driving most loudspeakers but takes


Fig. 7 : (above) The scratch fitter-c/assic 2nd. order conflguration.
Fig. 8 : (right) Scratch filter characteristic.


Oscillograms showing the smooth crossover for an output power of 100 mW r.m.s. into $8 \Omega$.
9a (above): shows 1 kHz triangular wave. Scales: X $200 \mu S / d i v$, Y $500 \mathrm{mV} / \mathrm{div}$.
$9 b$ (below): shows $10 k H z$ triangular wave. Scales: X $20 \mu$ Sidiv., Y 500 mV /div.


Fig. 10: Transient response driving 20W into $8 \Omega$ in paralle/ with $2 \mu$ F. Scales: X $200 \mathrm{HS} / \mathrm{div}$. Y $10 \mathrm{~V} / \mathrm{div}$.


Fig. 11 : Transient response when driving 1W into Quad electrostatic loudspeaker. Scales: X $200 \mu$ S/div., Y $2 V / d i v$.
care of the output under nearly no-load conditions. These arise when using headphones or with some loudspeaker crossover systems which look very inductive at higher frequencies.

## Headphone output

This facility was added at popular request and originally consisted of a jack socket straight across the output. An arrangement which promptly ruined a brand new pair of Koss PRO-4AA Stereo headphones not to mention a near miss on my eardrums. Although the former were kindly repaired free of charge, thereafter an attenuator seemed a good idea. The attenuator values were chosen to give a listening level very roughly equivalent to that through the loudspeakers and to be capable of driving normal low impedance headphones as those mentioned.

## Overload protection

This is another brain-teaser. It is very easy to get so carried away with protection circuitry that it becomes as complex as all the rest of the amplifier put together. Simple methods frequently introduce distortion and do not protect much anyway. In the Texan the only protection is provided by a fuse. It gives satisfactory protection against a short circuit on the output line since the TIP41A and TIP42A have a continuous collector rating of 6 amps which allows an adequate margin for the 2 amp fuse ( 1 amp for $15 \Omega$ loads) to blow.

One disadvantage of using a d.c. coupled output is that a fault in the output stage could put a large continuous current through the speaker. The fuse will also protect against gross overloads but this is, of course, rather dependent on the actual loudspeaker used. It is very unlikely that a catastrophic breakdown will occur but if belt and braces are required then the traditional large electrolytic, could be inserted providing the loudspeaker is returned to the -25 V supply line to polarise the capacitor.

The power supply for a class $B$ amplifier is the part most likely to be the subject of compromise. If size, weight and most of all, cost are of no consequence then a huge transformer and vast electrolytics will give almost perfect regulation. Alternatively, a regulated power supply can be built


Fig. 12 Circuit of the mains power supply for the "Texan" stereo amplifier. Note that a 2 A fuse (F2, 102) should be shown in the +25 V and -25V supply rails respectively.
which cuts down on the electrolytics but adds at least one extra power transistor. Unless carefully designed they can also suffer from reactive output impedances or worse still instability which does not help the amplifier, of course.

The circuit diagram of the compromise chosen is shown in Fig. 12. For the output stage nominal $\pm 25 \mathrm{~V}$ supplies are derived from an unregulated full wave circuit with $3500 \mu \mathrm{~F}$ smoothing capacitors. The 25 V rails in turn supply the $\pm 15 \mathrm{~V}$ rails for the operational amplifiers.

The 15 V supplies are stabilised by a pair of low cost zener diodes. Accuracy is not at all important here since the operational amplifiers will operate equally well with supplies between $\pm 12 \mathrm{~V}$ and $\pm 18 \mathrm{~V}$. The zeners are necessary more as clamps to ensure that the rail voltages do not exceed $\pm 18 \mathrm{~V}$, the maximum rating for the operational amplifiers. In parallel with the zeners are two $1000 \mu \mathrm{~F}$ capacitors, C20 and C21. These may seem rather high in value but with full bass boost and magnetic pickup equalization, an equivalent input in the order of $10 \mu \mathrm{~V}$ at 20 Hz will produce 20 W at the output. It is, therefore, essential to reduce ripple feedback to a minimum. The same reason, of course, accounts for the ease with which careless earthing can introduce low frequency instability.

The mains transformer recommended for the Texan is of rather unusual construction. It is a new design for which the manufacturers claim very low stray magnetic field-a great help when positioning the amplifier near a magnetic pick-up.

It also has a height of only $1 \cdot 5$ in. which gives the Texan its 'slim-line' dimensions.

The nominal secondary voltage is 20-0-20 at 1 amp continuous r.m.s. and measurements on the transformer in circuit showed a regulation of about 3 V r.m.s. between no load and a continuous sine power of 20 watts into one $8 \Omega$ load. The d.c. rails variations are as follows:

$$
\begin{array}{ll}
\text { Quiescent } & \pm 32 \mathrm{~V} \\
20 \mathrm{~W} \text { sine into } 8 \Omega \text { (one channel only) } & \pm 23 \mathrm{~V} \\
16 \mathrm{~W} \text { sine into } 8 \Omega
\end{array}
$$

(both channels simultaneously) $\pm 17 \mathrm{~V}$ It is apparent from these figures that the power supply is a major influence on maximum output power and distortion when considering continuous sine wave inputs.
Although holders for the integrated circuits were not used on the prototype, it is however recommended that they are used by the constructor.


## ISSUES WANTED

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# deluxe $9 / 2$ receiver 

 fortheSW, marz R.H.LONGDEN
## CRYSTAL FILTER

Fig. 6 is the circuit of this unit and if fitted is placed between the oscillator coil L3 (Fig. 2) and the i.f. amplifier described. The filter i.f. transformer provides a balanced output at pins 4 and $6, \mathrm{TCl}$ being adjusted to compensate for stray capacity in the crystal and parallel switch. In these circumstances selectivity can be extremely high but as the crystal works into a moderately high impedance here the passband is wide enough for a.m. reception without too much deterioration of speech quality.

When the switch is closed, the crystal is out of use, and the selectivity is that provided by the four i.f.t.'s.

The filter i.f.t. has to be modified, to obtain a centre-tap (pin 5 is unsuitable). Pull the screening can off the i.f.t. A small capacitor is wired internally between pins 4 and 6 and this component is snipped off. (The similar capacitor between pins 1 and 3 must be left). The i.f.t. is then replaced in its can. The parallel capacitance is now provided by the two capacitors Cl and C 2 , wired externally.


Construction. The filter is assembled complete on a piece of paxolin or Veroboard $3 \times 1 \mathrm{in}$. as in Fig. 7.

A black flying lead (or pin) is provided for the negative connection. The unit is mounted by two lin, bolts, with extra nuts which provide the chassis return. When the unit is fixed to the receiver chassis, the lead from pin 8 on the oscillator coil is near the filter i.f.t., and can be soldered to pin 2. The lead from pin 2 of i.f.t.l, (Fig. 4), is now soldered to the collector lead of the filter unit transistor or to a pin used as anchorage.

A hole must be drilled in the chassis to reach the lower core of the i.f.t. Two insulated leads are soldered to the crystal holder sockets passing directly down through holes to the crystal switch.

The crystal switch is fixed directly under the crystal. It can be a rotary switch, fitted to a bracket, with a coupling and extension spindle. Alternatively a switch wafer can be attached to the back chassis runner with ${ }_{2}$ in. spacers and operated by a ${ }_{4} \mathrm{in}$. dia. polystyrene rod cut to fit.

Filter Adjustment. A signal generator and output meter are helpful, though adjustment is possible by ear. The highest selectivity obtainable is very sharp so having adjustments a little off maximum is an advantage. Adjustment may need some patience, depending on how close the crystal and i.f.t. frequencies match at the beginning.

If alignment is badly out, no signals at all may be heard. If so, align the i.f.t.'s as described earlier, switch the crystal out (switch closed) and adjust the two filter cores for best volume. Set the phasing capacitor TCl about half closed.

Fig. 6 : (top left) shows the circuit for the optional crystal filter, IFT1 being modified as described in the text. Fig. 7: (left) gives the layout of the filter on its $3 \times 1 \mathrm{in}$. board.


Fig. 8: Curcuit of the audio stages, fed from the detector stage (Fig. 4, Part 1) via the volume control VR2 mounted on the front panel.

Open the crystal switch, and tune the receiver (or generator). If results are much as before, crystal and i.f.t.'s are by no means on the same frequency. Assuming no generator or other means of checking the frequency of the i.f.t.'s is available, switch the crystal out, adjust all seven i.f.t. cores about a half turn in the same direction, and re-peak them for best results. If bringing in the crystal now sharpens tuning, the shift made to the intermediate frequency was correct. If results seem little changed or worse, the shift was the wrong way, so all the cores have to be moved in the opposite direction.
When crystal and i.f.t.'s are approximately in agreement, tuning through a signal (or tuning the generator) will show two tuning points. One is extremely sharp, and is not changed by adjusting the cores of the i.f.t.'s in later stages. The other is flatter, and can be moved by rotating the cores. The sharp peak is the crystal resonance. Tune to it and then rotate each i.f.t. core to the same frequency, as indicated by a great increase in sensitivity and the eventual disappearance of the second, flatter response.

With i.f.t.'s aligned to the crystal frequency, tune through a signal (or tune the generator) while

adjusting the phasing capacitor TC1. If TC1 is completely unscrewed, a very sharp null will be found at one side of the signal, the signal re-appearing as this is passed. If TCl is fully screwed down, the null arises at the other side of the signal, so it is adjusted to give a reasonably even response. Adjustments may make necessary very slight readjustment of the filter cores.

## AUDIO AMPLIFIER

With an unlucky combination of values in the capacitors C 1 and C 2 the error could prevent phasing. If so, place a 30 pF pre-set across C 1 or C 2 , re-adjust the core, and adjust as described. It may be found that exchanging C 1 for C 2 will suffice. If selectivity worsens with the pre-set at any value, remove it and put it across the other 470 pF capacitor.

The circuit for this is shown in Fig. 8. It requires a few more components than directly-coupled circuit but has the advantage that d.c. operating conditions for one stage do not depend on those of other stages, so that unsatisfactory working is not likely, while all items are easily obtained.

Construction. The parts are assembled on an insulated panel ${ }_{31}{ }_{2} \times 2 \mathrm{in}$. as in Fig. 9. The amplifier can be teasily tested by adding a speaker, 9 V battery and providing some form of input, such as a pickup, or by connecting it to the receiver volume control.

The amplifier is mounted by two $1_{2}$ in. 6BA bolts, with extra nuts, which also secure tags, and act as a return to the receiver chassis. In the event of the amplifier drawing a very small current with no signal (say under 3 mA ) and exhibiting cross-over distortion, the $82 \Omega$ resisfor R10 may be increased slightly in value, $100 \Omega$ being suggested.

Fig. 9: Layout of the audlo ampl/fler constructed on a $3 \frac{1}{2} \times 2 / n$. insulated board. It is mounted on top of the chassis with 72 at the rear, see Fig. 12 and photographs.


Fig. 10: Circuit of the b.f.o./product detector required for the reception of c.w. and s.s.b. The function switch S1,2,3 is mounted on the front panel.

## BFO/PRODUCT DETECTOR

The receiver as described so far is intended for a.m. reception and is thus suitable for all normal short wave listening. In some cases it may be wished to receive Morse and single sideband transmissions, especially on the amateur bands. This can be done by using a beat-frequency oscillator/carrier oscillator, with product detector.

The circuit for this unit is shown in Fig. 10. Trl is the oscillator, operating at 1.6 MHz . This frequency is varied by the panel pitch control VCl. The zener diode ZD1 provides a stable voltage for the oscillator.

Output from the oscillator goes to one of the pair $\operatorname{Tr} 2 / \operatorname{Tr} 3$, with c.w. or s.s.b. signals fed in from the i.f. amplifier via C8.

For c.w. reception, VCl is adjusted so that the required audio heterodyne note is produced.

For s.s.b. reception, $\operatorname{Tr} 1$ operates as carrier oscillator, to replace the carrier removed during s.s.b. transmission, thus allowing the s.s.b. signal to be resolved.

A rotary function switch provides for c.w./s.s.b. reception or a.m. Section S1 interrupts the negative line for a.m. reception, while section $S 2$ switches the volume control VR2 to the product detector circuit, or existing a.m. diode detector, as required. Section S3 brings in R12 for c.w./s.s.b. to complete the diode circuit made through VR2 during a.m. reception.

Construction. The components are assembled on an insulated board as in Fig. 11. All components, including the b.f.o. coil, are on this side of the board, so it must be checked that coil connections are correct when turning the board over.

Flying leads are provided for external connections and the board mounted clear of the chassis by $1_{2}$ in. bolts with extra nuts. The point MC is a tag locked to the board thus completing the circuit to the metal chassis when the assembly is mounted in place.

Capacitor C8 is near the final i.f.t. a lead being soldered to pin 4 of the latter and passing through the chassis to C .

A black lead from R4 is connected to S1, which completes this circuit for c.w./s.s.b. reception only.

VR2 is disconnected from the diode circuit and taken to S2. This connects to C 7 in the c.w./s.s.b. position and to the diode (as originally) for a.m. reception. R12 is wired directly to the switch as in Fig. 10.

Alignment. With the switch in the a.m. position, reception should be as before but the final i.f.t. core will require a slight adjustment for best results.

Set VCl half closed, switch to c.w./s.s.b., and rotate the b.f.o. coil core until a strong heterodyne is heard. Leave the core at the central 'zero beat' position. Turning VC1 either way will now produce a heterodyne which rises in pitch.

Fig. 11: Layout of the b.f.o./p.d, which is mounted under the chassis below the audio amplifier board.

## CRYSTAL FILTER

Resistors:
R1 $1 \mathrm{k} \Omega \quad \mathrm{R} 2 \quad 22 \mathrm{k} \Omega \quad \mathrm{R} 3 \quad 4.7 \mathrm{k} \Omega \quad \mathrm{R} 4 \cdot 1 \mathrm{k} \Omega$
All 2 W $10 \%$
Capacitors:
C1 470pF $1 \%$ silver mica
C3. $0.1 \mu \mathrm{~F}$
C2 $470 \mathrm{pF} 1 \%$ silver mica
C4 $0.1 \mu \mathrm{~F}$
TC1 2-10pF trimmer (Jackson 'Tetfer')

Miscellaneous:
Transistor Tr1. AF117. IFT1 (Denco IFT18/1 6 see text for modification), Crystal $16 . \mathrm{MHz}$ type HC6U (Senator Crystals) and holder. S1, Rotary on-off switch with extension spindle and couplingh (see text). Veroboard, plain, $3 x$ in:

## AF AMPLIFIER

## Resistors:

| R1 | $3 \cdot 3 \mathrm{k} \Omega$ | R5 | $8.2 \mathrm{k} \Omega$ | R9 470\% | R13 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R2 | $120 \mathrm{k} \Omega$ | R6 | $82 \mathrm{k} \Omega$ | R10,82, |  |
| R3 | $47 \mathrm{k} \Omega$ | R7. | $22 \mathrm{k} \Omega$ | R11 $10 \mathrm{k} \Omega$ |  |
| R4 | $12 \mathrm{k} \Omega$ | R8 | $680 \Omega$ | R12 10ks |  |
|  | W 1 |  | -R10, | and $125 \%$ |  |

Capacitors:
$\begin{array}{ll}\text { C1 } 2 \mu \mathrm{~F} 6 \mathrm{~V} \\ \mathrm{C} 2 & 50 \mu \mathrm{~F} 6 \mathrm{~V}\end{array}$
C3 $6 \mu \mathrm{~F}$ 6X
C4 $50 \mu \mathrm{FBV}$

## Semiconductors:

Tr1 OC71 Tr2OC81D TS 4 OC81

## Miscollaneous:

T1, driver transformer (Home Radlo TR64), Y2: output transformer (Home Radlo TR65A). Yeroboard plain, $3 \frac{1}{2} \times 2 i n$.

## BFOIPRODUCT DETECTOR

Resistors:

|  | 22k |
| :---: | :---: |
| R2 | $4.7 \mathrm{k} \Omega$ |
| R3 | $1 \mathrm{k} \Omega$ |
| R4 | $470 \Omega$ |
|  | IW 10 |



## Capacitors:

C1 $0.04 \mu \mathrm{~F}$
C4 18pFSM C7 $4 \mu$ FIRV C5 $100 \mu \mathrm{~F}$ 12V C8 , 47pF SM C6. 0.024F

VC1 15pF váriahtik (Uackson Type C804)
Semiconductort?
Tr1 OC45 Tr2 OC46 Tr3 OC45 ZD1 zener diade 5.6 V

## Miscellaneoius:

BFO (Denco IFT 17/1.6), $81-3,3$ pole 2 way rotary switch. Veroboard, plain $31 \times 2410$.


Fig. 12: Location of four of the units and the battery, the b.f.o./p.d, belng mounted underneath. Compare with photographs. An aluminium screen was later fitted between $L 1$ and $L 2$ on the r.f. unit, to improve stability. This can be seen in the photographs. Mounting instructions are included with the tuning dial.

When a c.w. signal is tuned in adjust VCI for best reception the b.f.o. being above or below the signal, whichever results in least interference from other signals.

When receiving s.s.b., VC1 is carefully adjusted to give best resolution of speech. If resolution is impossible, the oscillator is being tuned to the wrong side band of the s.s.b. signal. The same sideband is nor-
mally in use on any one amateur band, so the adjustment of VC1 will hold good for that band.

With s.s.b. it is normally necessary to reduce r.f. gain somewhat by operating VR1. If the crystal is in use, tune in the s.s.b. with the switch at a.m., then switch to s.s.b. and adjust VC1 for best resolution. VCl can then be generally left untouched while tuning the band for other s.s.b. signals.

## GENERAL CONSTRUCTION

Connections and placement of controls, etc., are as follows:
VR1. This is fitted at the right of the panel. Leads from the r.f. and i.f. units go to one outer tag. The centre and other outer tags are wired to chassis.

VR2. This fits at the left. ZD1 positive runs to one outer tag, the centre tag goes to the a.f. amplifier, and the other outer tag to chassis. (If rotating the


Panef layout. Top left, Volume control and top right, r.f. gain control. Left to right below:-b.f.o. tuning, function switch, crystal filter in/out, mixer and aerial tuning.
knob gives reversed control of volume, change over the leads to the outer tags. This also applies to VR1.) Connect one switch tag to chassis, and a red lead with positive clip to the other switch tag.

VC4. This is located between the crystal switch and VR1, and is connected to a lead from VC1. VC5 is similarly connected to VC2. Rotor tags are earthed to an adjacent tag bolted to the chassis.

Negatives. An insulated tag is bolted underneath the chassis near the a.f. amplifier, and is an anchor point for battery negative, and the negative leads running from each unit.

Aerial-Earth. Two sockets are fitted to the rear chassis runner. That for earth is connected to the chassis. The lead from pin 8, L1, runs to the Aerial socket.
Speaker. The secondary leads from T2 run to separate sockets, or a jack outlet. A $2 / 3$ ohm speaker is required, fitted in a cabinet or attached to a baffle board. For c.w. reception low impedance headphones should be used.

Panel. This is secured to the chassis by the control bush nuts and two brackets. The tuning drive is fitted by the method described in the instructions supplied with it.

Cabinet. Holes are punched in the back, level with the sockets. An opening lid is made in order to reach the coils and the battery, which rests on the chassis near the a.f. amplifier. Small holes are drilled close together in order to start a metal-saw, and a piece about $10 \times 5$ in. cut out. Two strips of metal about $6 \times 1$ in. are fixed inside with countersunk screws, for the lid to close against. It swings on a 10 in , hinge bolted at the back.

## CO! CO!CO!CQ!CO!

## IMFORMATION WANTED

.can anyone suggest a suitable spaaker for use with my 'Mellophone 3' set. It is mounted in a smart mahogany cabinet standinq on legs. Valves used are PM2DX, is mounted in smart mahogany cabinet standing on regs. valves used are PM20X, Lancaahlre.
...can anvone please lend me a copy of Modern Wireless No. 1.-L. A. Slack, 60 Parsonage Road, Withington, Manchester, M20 9 WQ.
.1 have recently acquired a Carlton SG 4 receiver about 40 years old and 1 cannot find any gen on it at all (vaives, typess, etc.). Please can anyone help.-A. C. Lintott 3 Seaham Street, Silksworth, Sunderland, SR3 1EX.
details and copies of radio circuits from early editions of radio magazines, as I wish to renovate some Vintage receivers.-S. Laffan, St. Kieran's College, Kikenny Ireland.
BSitupW.
...any gen on the Rx type R1475 p.8.u. type R360 and tuner unit type 131.-C. Astbury 16 Eaton Avenue, Handbridge, Chester, CH4 7HB.
any gen on R4187 S.t.C. set 17 valve, 24 channel receiver circuit. Manual and mods for malns. Loan or buy. Also back nos. P.W. for last 10 years. Sell at cover price plut postage.-K. Armsirong, 11 Woodvilie Gardens, Rulsip, Middx.
...circuit and details of R209 Rx. Also gen on 82 (spy) receiver and layout if possible. i. L. Blehop, 27 Stratton Way, Blggleswade, Beds.
…any gen on Reiner Electronics Oscilloscope No. 556. Service Manual, modification detalls etc. Will return.-R. P. Davies, 61 Kenley Road, Kingston, Surrey.
...circuit or Information on capacitor analyser (Solar) type CB-2U. Loan or purchase. -i. Kirk, 22 Barncraig Street, Buckhaven, Fife, Scotland.
.icircuit diagram and gen on the R1155 Rx. I wlil copy and return.-A. McKenzle. 6 Moorfield, Newcastle upon tyne, 2 .
.i.service sheet for Pamphonic, Reproducers Lid. type 3000 stereo amplifler. R. W. Parker, 11 Greenacres, Downton, Sallsbury, Wilts.
...any gen on "Contessa" recelver which Is 5 to 8 years old with 6 transistors and a diode and what appears to be 2.l. stages. The tuning dial coniains the word Contessa and at the bottom of the speaker fret it reads "transistor"-H, Gosing, 12 Uilswater Avenue, Halfway, Sheffieid.
, .any info on making a rumble and scratch fitter.-N. Davidson. Leamington House, any hints and mod, Kent.
nited World College the p.s.u. of the 19 set 80 that we can remove dynamotors.CFs 9WF.
i...Insfruction manual for Airmec 877 Televet test set.-D. Wigley, 4 Sollons Avenue, Harlesden, London, N.W. 10.
Wood Lany info on converting the 19
 161 Octavla Court, Greenock, Renfrewshite, Scotiand.
10 Nircuit and any Info on constructing an echo (depth) sounder.-R. Mediand, 10 Northfleld Way, Brighton, BN1 BEH.
..blueprint or practical circuit for the 4-fransistor amplifier published in P.W November 1961-B. Endean, 25 Astley Gardens, Seaton Sluice, Whitley Bay. Northumberland.
..circuit diagram of an electronic echo chamber: the type which does not employ an endless tape with multiple heads but one which will produce both sustalned echo reverb) and repetitive echo with a variable time factor.-M. Hale, 27 Rachel Gardens. Birmingham, B 296 NH
...the circuit diagram of a stylus organ or even the details of the one published in P.W. some years ago.-V. P. Kelly, L/103197 REM(A) Mess 6F1, H.M.S. Ark Royal BFPO, London.
...blueprinfs for the following: "Celeste 7-transistor portable" (June 1963), "7Mc/s Transcelver" (June 1984), "8eginners S.W. Superhet", (December 1964) Progressive P.O. Box 2671, Tripoli, Libya.
 weicome.-D. Millard, 20 Jubilee Road, Tipton, Staff, DY400P

## EQUIPMENT REQUIRED

$\mathrm{H}^{30}$ urce of supply of small ferrite-cored r. 4 . chokes of $680 \mu \mathrm{H}, 470 \mu \mathrm{H}$ and $330 \mu \mathrm{H}$. -M. J. Shepherd. 72 Westerland Avenue, Canvey Island, Essex, SS8 8JS.
102 Where can obtain 1.t. transformers for a Skyrover Radio.-J. B. Valentine. 102 Woodhouse Lane, Springfeld. Wlgan, Lancs.
photocell fine line scanner and smplifier to work quick-acting solenoid for llghtwelght equipment. Black and white, on and off.-D. Corbett, 41 Tinto Hoad, Gearsden, by Glaegow.

## BOOKS WANTED

coples of the Marconigraph, or any other old radio mags. pre 1930 . Also a nice ot of Harmsworth's Wireless Encyclopedia.-C. Riches, Practical Wireless Editorial Dept., Fleetway House, Farringdon Street, London, EC4A 4AD.

## CORRESPONDENTS WANTED

.any one interested in electronics whether it be ham radio. Hi-Fi. audlo, TV o any other aspect,-Des. Walsh, E15CD, "Coombe Down", Ballylynch, Carrick-on Sulr, Co. Tipparary, Ireland.
..P.E. May and July 1970. Will exchange for P.E. September to October 1969.-
Des. Walsh, "Coombe Down". Ballyynch, Carrick-on-Suir, Co. Tipperary, Ireland

## I8SUES WANTED

1950-1955 and Radio Const. 1950-1959,-Lurton, 12 The Vale, Acton, London. w3 7SB.

Feb. and March 1971 P.W.-R. Wateon. 51 Wocdbrldge Road, Newbourne, Nr. Woodbridge, Suffolk.

Television for Oct, 1967, Feb.., March, April, May, June, July 1968; Aprif, May, June, July 1969 and Dec. 1970.-P.F. Johnson, 21 Jubilee Drive, Sheringham, Norfolk and Dec. 1954-i. Simpson, 2 Taylor Cottages, Derryhale, Portadown, Co. Armagh. N.I.
N.i.Jan. to Nov, 1966, July 1967, Jan., Feb, March 1968, Index for Vols. 41 to 46 (all P.W.), P. Electronics for Jan to Nov. 1966, Nov, Dec. 1967, Jan., March 1968, Jan. to Nov. 1989, Nov. 1970, Aprll 1971 and index for Vois. 1 to 6.--H. Ferns, 11 King Street, Coatbrldge, Lanarkshlre, Scottand,
...lssue of P.W. Jan. 1962 containing 'MInuette".-S. Grosvenor, 50 Countess Crescent, Bispham, Blackpool, FY2 9LO
P.W. for June and July 1969 also Jan. 1971. P.E. issues Jan. to March 1970 inclusive.-A. Taylor, 89 Howth Drive, Woodley, Nr. Reading.
..P.W. September and Oetober 1970.-W. R. Pitt, 26 Highdown Avenue, Worthing. Sussex.
P.W. June 1971. F. J. Camm's "Beginners' Guide to Radio" (5th or 6th edition) Also "Eiementary Course in 28 Lessons", and P.W. July 1948.-W. Howard Shepherd, 5 Pafford Avenue, Watcombe, Torquay, $\mathcal{Y} Q 2$ 8BS.
-E. Farrer, 108 Colne Road, Twickenham, Middx.
Flat.W. Sept., Oct, and Dec. 1970 and Oct., Nov, and Dec. 1971,-C. J. Page ..A April Devitinshire Terrace, Lancaster Gate, London, W.2.

Apr. J. July issue of P.W. $\rightarrow$ T. M. Fresson, Fernilee, Churt, Farnham, Surrey.
P.W. Auq 1960-T. Richards, 52 Cookson Street, Liverpool, L1 5EH.

10 Mayflower Road Shirley South containing the Portable Keyless Organ.-K. Oliver, 10 Mayflower Road, Shirley, Southampton, Hants
P. T. Jyly 1971 with the 625 i.f. strlp.-O. Collins, Landiord, Germansweek, Bear-
worthy, Devon.

## TAK <br> AKE <br> 2®

 JULIAN ANOERSON> A series of simple transistor projects, each using less than twenty components and costing less than one pound to build.

THIS month's project is one for the DXers; it is a noise limiter circuit that can either be built into a receiver or used as a separate unit, in which case it fits between the headphones and the receiver.
Noise is a real curse when listening to weak, far off stations and of course its amplitude bears no relationship to that of the signal that you want to hear. A comparatively simple circuit, Fig. 1, will eliminate all signals above a certain level while unaffecting those below that level. Our ears have a sort of a.g.c. system built into them, if they are subjected to a series of loud noises they will not be very sensitive to quieter passages in between those bursts, so, however loudly you turn up the volume, this will not improve intelligibility. By limiting the level of the peaks to that of the wanted signal there is a considerable improvement and signals which were quite lost in the noise can be heard comparatively well.

Let us assume that the unit is to be used from the output of the receiver. This output may be the equivalent of an external loudspeaker socket or a special headphone output, it depends on the design and type of receiver. If the socket is for an external Ioudspeaker and the output stage is a transistorised one, it is important that the output load is maintained and a resistor equivalent to the speaker impedance should be fitted, in the circuit shown as R1. If you are not sure what the value should be $10 \Omega$ will be satisfactory.
The input is connected via the level control VRi to Trl where the signal is amplified. The output of this is fed via C3 to the back-to-back silicon diodes D1 and D2. These are not directly biased but as soon as the level of the signal across these goes over about 0.6 V , they will conduct; one of them clips the positive peaks while the other takes care of the negative peaks. It doesn't matter how high the signal level is, anything above this level will appear as a maximum of 0.6 V . Since we shall now always have the same level at this point we have removed our control over the listening volume and for this reason VR2 is included.

If the wanted signal level is above the critical 0.6 V , this will introduce distortion and it is in this way that the controls are set. The receiver's volume control should be set for normal listening level and, with the unit connected, VR1 should be increased until the wanted signal becomes distorted and harsh; this should then be backed off slightly. VR2 is then set for a comfortable listening level.

The current consumption of the circuit is not high and a small PP3 battery will cope nicely. For fixing the circuit permanently into a set, VR1 can be the existing volume control, it doesn't matter even if the value of this is considerably higher than that

## No. 36 NOISE LIMITER



Fig. 1 (above): The circuit


## $\star$ components list

| R1 | 102, 5W-see text | - |
| :---: | :---: | :---: |
| R2 | $1.2 \mathrm{M} \Omega 5 \%$, tW | 1p |
| R3 | $2 \cdot 2 \mathrm{k} \Omega 5 \%$, t W | 1 p |
| C1 | $5 \mu \mathrm{~F} 10 \mathrm{~V}$ minimum | 4 p |
| C2 | $50 \mu \mathrm{~F} 10 \mathrm{~V}$ minimum | 4 p |
| C3 | $5 \mu \mathrm{~F} 10 \mathrm{~V}$ minimum | 4 p |
| Tr1 | BC109 | 12p |
| D1 | 1N914 | 5p |
| D2 | 1 N 914 | 5p |
| VR1 | 10kS log. pot | 12p |
| VR2 | 10k $\Omega$ log. pot | 12p |
| SW1 | On-off switch | 10p |

Prices are those recently advertised in Practical Wireless and may have changed. No allowance is made for minimum order costs or for postage and packing and these should be checked carefully before ordering.
shown. The output from VR2 can then be connected to the point that was previously fed from the volume control's slider. If used with a valve set the supply voltage can be taken from the h.t. line via a suitable dropping resistor to give 9V. A switch can also be included to switch the unit either in or out.


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# THE BROADCAST BANDS Malcolm Connah Frequencies in kHz - Times in GMT 

## MONTHLY NEWS FOR DX LISTENERS

THE recent spell of cold weather seems to have kept most of the readers of this column in their shacks and at their writing desks, this is evidenced by the number of reports received-the record having been beaten yet again.

The first report comes from Ian Howes in Lowestoft who used his R209 Mk II receiver and TV antenna to log the following interesting stations:-

3375 AIR, Gauhati with Indian Music at 1700.
4780 R. Carabobo, Venezuela, Spanish at 0015.
4860 AIR, Delhi, Domestic Service at 1800.
4880 R. Universo, Venezuela, Spanish at 2315.
4965 SABC, Domestic Service at 2100.
5000 R. Nepal in Nepali at 1540.
5035 Alma Ata, USSR at 1600.
5035 Bangui, Cent. Afr. Rep., French at 2230.
5047 R. Lome, Togo, news in French at 2200.
5052 R. Singapore, English Pops at 1600.
9660 ABC, Brisbane, Home Service at 1200.
11835 R. El Espectador, Uruguay, Spanish at 2304.
15430 Voice of Free Korea in English at 1130.
D. A. Hairon of St. Clement, Jersey has again used his Codar CR70A to good effect hearing: -

4915 R. Accra, Ghana in English at 2300.
6170 Israel B.C. in English at 2115.
9505 R. Record, Brazil in Portuguese at 0110.
9510 R. Barquisimeto, Venezuela, Spanish at 0035.
9625 BBC, Cyprus relay in Hindi at 0050.
11775 AIR in English at 1050.
11850 R. Accra, Ghana in English at 2030.
11875 R. Soc. de Bahia, Brazil at 0000.
15084 R. Tehran in Farsi at 1430.
15105 BBC, Ascension Is. relay at 0730.
15225 R. Cultural de Bahia, Brazil at 0100.
15230 RT Belgium in English at 0050.
17825 NHK, Japan in English at 0815.
Michael Berry of Dewsbury, Yorkshire is a new reporter to this column, his equipment consists of an Eddystone EB35 receiver and a 15 foot vertical antenna. Amongst the interesting stations heard were:-

4870 Cotonou, Dahomy in French at 2230.
4915 R. Accra, Ghana in vernacular at 2215.
4940 R. Abidjan, Ivory Coast in French at 2305.
6040 Tashkent, USSR in English at 1420.
9520 R. New Zealand in English at 0900.
9570 R. Kaduna, Nigeria in English at 2100.
9570 R. Portales, Chile in Spanish at 0005.
9850 R. Bangladesh, Dacca in English at 1515.
11710 R.A.E., Argentina in English at 2335.
11735 RTV Morocco in French at 1910.
11805 R. Globo, Brazil in Portuguese at 2105.
11865 R. Kinshasa in French at 2230.
11925 Tashkent, USSR in English at 1410.

11925 R. Bandierantes, Brazil in Portuguese, 2145.
15190 ETLF, Addis Ababa in Arabic at 1430.
15520 R. Bangladesh, Dacca in English at 1230.
Julian V. Moss of Rayleigh used his Meridian 10 transistor superhet and 60 foot long-wire antenna to compile his latest log which includes:-

9515 R. Ankara, Turkey in English at 2200.
9525 AIR, Delhi in English at 1945.
9545 R. Accra, Ghana news in English at 2045.
9605 R. Canada International, English at 2100.
9695 RSA, South Africa in French at 2015.
11620 AIR, Delhi news in English at 1945.
11710 VOA, Liberia sign off at 2230.
11720 R. Canada International, English at 2207.
11780 HCJB, Quito, Ecuador in English at 2000.
11850 R. Accra, Ghana news in English at 2030.
11900 RSA, South Africa in English at 2235.
11910 ETLF, Addis Ababa in English at 2015.
11920 Abidjan, Ivory Coast in English at 1845.
11950 FEBA, Seychelles in English at 1730.
11970 RSA, South Africa in English at 2240.
Brian Ewing of Ilford is another reporter with a Codar CR70A receiver, this time used with a 50 foot long-wire antenna to hear:-

6010 RAI, Rome, Italy at 0110.
6155 R. Renascena, Portugal (tentative) at 0030.
9510 BBC, Ascension Is. relay at 0415.
9615 HCJB, Quito, Ecuador at 0330.
9695 RSA, South Africa noted at 221.5.
9730 ETLF, Ethiopia at 0400.
11815 TWR, Bonaire noted at 0130 .
11850 BBC, Tebrau, Malaysia relay at 2345.
11925 Radio Kuwait at 2040.
11930 WIBS, Windward Islands at 2031.
11955 Radio Lebanon at 0235.
15170 ELW A, Liberia at 2159.
The last reporter for this month is Fred Wall of Walthamstow who has a Plessey PR155 (short pause whilst your scribe goes green with jealousy) and a 20 foot long-wire enabling him to hear:-

9460 Radio Pakistan with music at 2020.
9520 Radio New Zealand in English at 0900.
9745 Radio Baghdad with music at 2040.
9755 Radio New Zealand at 1700.
15240 Radio Australia with news at 2153.
15245 RTV Morocco, news in Spanish at 1415.
15345 Radio Kuwait, Pop Music at 1600.
17820 Radio Ankara, news in English at 1330.

Reports should arrive by the 15 th of the month and be addressed to me at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.


## THE AMATEUR BANDS David Gibson, G3JDG

## Frequencics in khlz • Times in GMT

THINGS are hotting up r.f.-wise at Moretons, Harrow. The "Dorms" are bedecked with sneaky antennas of varying shapes and sizes and are used by an enthusiastic band of s.w.ls. Jonathan Thackeray is one and pleads guilty to listening on the Amateur bands by fixing an H.A.C. one-valver onto his modest 15 ft . end fed. If you are having a chuckle at all this, observe Master Thackeray's log for 80 metres; JA5TY, JY1/B, UL7IAF, UP2TP, VE1FO, VE3QA, VK2LF, VO1MG, VP9UN, VQ9R, 4X4DP, 9H3B, 9K2AL. Definitely ten out of ten for that splendid effort. (Write out one hundred lines, "I must not receive better DX on eighty than G3JDG!")

It seems that most listeners kept their ears draped over the l.f. bands this past month. Log from David Lawley (Gravesend) shows a fair bit of activity on top band. Gear is a CR7OA plus 220ft. folded longwire with a quarter wave counterpoise earth for topband. All stations are c.w. unless marked; DJ4SS (s.s.b.), DL1CF, DL9KR, EI8BF, EI9BG, EI9J, GD3GM (s.s.b.), GM3-IGW, UKG, YCB, YOR, ZDH, GW3-UCB, UPK, YTO, HB9-CM, NL, OK1-ATP, AYY, KYS, MA, SBT, OL1AOH, OL5ANJ, PA0PN, W1BB/1, W1HGT. David bagged 195 countries last year, has already heard 85 this year and hopes to take the R.A.E. on May 9 (Gd lk OM).

John Runchman reports "A" level QRM but has managed a few late night/early morning sessions as a nice change from delta $y$ tending to zero! Receiver is a homebrew two-valve t.r.f. (ECC83/ ECL82-what funny transistor numbers) with a 20 metre dipole. Topband c.w. brought in: EI8H, HB9ANW, HB9CM, HB9NL, OK1ATP, OK1MAC, PAOPN. Just in case you think that two valves isn't very much you are quite right. The following log received on this set for 14 MHz tends to add weight to the argument that it's what's between the headphones that counts; CN8GG, EA6BM/M, EA8HR, G3RFP/MM, LU2AJE, PJ2CL, PY1CBS, PY1DG (no relation!), PY1ZAN, PY2DVH, PY4AP, PY7OS, PZ1DR, UC2DX, VO1CU, VP2MZ, VP9CP, YV4AOW, YV6IP, 4S7SW (Ceylon), 4X4BL, 7X2BK, 9G1FF, 9H1CQ, 9Q5ITU, 9Q5RD. All this with a t.r.f., that should get some "reaction" from the superhet fraternity? Incidentally, John's QTH is South Norwood, London.

News from Karl Muller in the land of 3D6 (that's Swaziland in Southern Africa) that FL8HM and FL8MM both operate on band edge of 7 MHz c.w. but he doesn't say the QTR. Karl has a Barlow-Wadley XCR-30 receiver with a 30 ft . end fed and an a.t.u. The following made it into Swaziland on 7 MHz c.w.; CM3LN, CR6AI, DK1EB, DU1IOR, FB8XX, FL8HM, FL8MM, HC2HM, JA1OHU, JA3EA, JA3PMB/6, JA5ADR, KP4DJE, KR6AY, LA2QI, OK2KYI, PY1BSN, PY3MU, TU2BK, VP2AAA, W6GRU, W7DI, ZD8BR, 4S7DA, 5R8BD, 6W8AL, 9 M 2 CW . Incidentally, on ten metres Karl logged G3UGM.

He also asks if people who work s.s.b. have callsigns and enquires why certain YU and LZ stations spend half an hour calling CQ on top of DX stations? (Yeah, well...).

Patrick Funnell reports on a pirate station on topband signing G4NCC who, apparently, spends his hours taking the mickey out of Amateur radio. Be warned friend, in this transistor age many Amateurs have some very simple but excellent d.f. equipment. All it needs is two of them to hear you and get a fix, and where the lines cross-that's you. Patrick has a BC312 receiver and a ". . . piece of wire in the loft." This set-up logged these on topband: EI9J, GM3IGW, GM3YOR, GM4AGG, OK1ATP, OK1JAX, OK1KYS, OKINRX, all on c.w. and heard between 1830 kHz and 1840 kHz .

Eighty metres seems to be a strange band. Some acclaim it as just great while others give a clear "thumbs down" sign, or something.

Steve Taylor (Abbey Wood, S.E.2) reckons there's a lot of "G" activity about judging by what he heard on his CR70A. Stefan Kaye (Witney, Oxon) sends in a good log of eighty metre happenings bagged with the aid of an AR88D and 180ft. longwire which is untuned. Pickings include: EA8HA, OA4OS, VE1ADV, VE1AU, VP2LAT, WB0FFG/TF, XEIKB, ZL3GS, ZL3RK, ZL4AV, ZL4KE, 4X4UF, 9H1BX, 9HIC, 9H3B.
"My first $\log$ to the Amateur Bands", says John Vaarnela of Croydon Park, New South Wales, Australia. Receiver is a BC348Q and the antenna a dipole at 25 ft . Squeaks of s.s.b. on 14 MHz heard from: A2CAB, CE1LU, CN2CG, CP1DN, CR6GA, CR7GJ, CX4RZ, EA8GZ, ELOK/MM, ET3JH,, FB8ZZ, FB0WK, FK8ID, FL8MM, FO8DO, FO0TZ, FR7AB, HI8LC, HK1CKY, HK0BKX, HR1BK, HS1ADR, IC8CQS (Capri), JD1KAA, KC6QS, KG6SI, KH6AQ, KR6TQ, KP4DIW, KS6DH, KV4AA, KX6PQ, KZ5GB, LU1SE, MP4BBW, OA4PQ, PJ2MI, PY3ATW, TI2RT, VP1BH, VP9AT, VQ9NEW, VR1AB, VR4EE, VR5FX, VS5PW, VS6BS, VS9MT, VU2ED, XE1KB, XW8DO, YA1AG, YB0AT, YJ8BJ, YK1AA, YN1HC, YS1DET, YV6JB, ZD6HJ, ZD8RR, ZE2KV, ZP5KU, ZS2DH, 3B8CJ, 3D2FM, 4S7PB, 4WIAF, 4X4YT, $5 H 3 L V, 5 N 2 A B G, 5 R 8 A B, 5 V 3 R U$, 5W1AU, 5X5NA, 5Z4KZ, 9H1ST, 9K2CP, 9M2GV, 9M8FMF, 9N1MM, 9V1VR. Who said twenty was dead?

Julian Iredale (Llandudno), JR500SE, Codar PR30, 264 ft . end fed got these goodies on 28 MHz s.s.b.: K4AUV, K4HGG, VE40CK, W2BLV, W4GJO, W8QUO, W0ZRX/MM ( 500 miles north of the Azores), WA5MAK, WB4BMV, WB5CTE, WB4WLI/ YV1, ZC4GLW, 4Z4BG, 5R8AP, 7Q7AA, 9J2DG.

[^5]
# practically wireless commentar by HLINTI 

SLAVE to the whims and wishes of my readersthe whims of one and the wishes of the other, did I hear you say?-Henry sits here in the light of a flickering candle, like a Dickensian P.W. Editor.

Henry has been remiss. Not gifted with foresight, he to whom tea-leaves are a greater mystery than Rorsach blots failed to provide auxiliary illumination against the day when the power should be cut. The bulb's gone in me torch, innit?

I should have invented something electronic, on the spur of the moment, say the cold eyes around the blank TV. Like all great discoveries, the answer should have come in a blinding flash-well, perhaps that is not the best simile. . .

As for example, when the transistor was born? Ah, but I'll bet it wasn't at all like that.
Shockley: Let's give it up. Nearly lunchtime.
Brattain: One more try. (He's the meticulous one. Like Spencer Tracy as Edison, he will not give up.)
S. I'm hungry. We are not going to get AFN without a tuning coil, and I'm sick of sharing the same bit of germanium.
B. Right. I'll just leave my catswhisker here. You can do what you like with yours.


A Dickensian P.W. Editor

## Shockley does.

They return, replete and burping. That Bell Laboratories Inc. canteen was notoriously shepherdpie oriented.
B. Next step, then.
S. Oh., let's start again.
B. No, you measure, I'll inscribe.
S. Just a minute while I suck a-hey-what's this?
B. What's what?
S. Amplification. . . .
B. Well, it is a process of-
S. No, lout. Look! 1 mV in, 100 mV out. You fiddled the meters.
B. I did not, indeed!
S. Well, somebody did. Let's pack up and invent a fuel cell.
B.-and twenty makes four-hundred-no, listen Shock, we've got something here. We can transfer effective resistance over a link circuit. We can leap a barrier.
S. Don't get romantic. Have some bicarb.
B. Bicarb, hell! Don't you see, this is a breakthrough. No more pottwiddling. Just stick a couple of cats' whiskers on a slab of germanium and the input resistance changes to a completely different output resistance just by an alteration of voltage. . .

Of course, dear reader, it did not turn out like that. Some fussy, officious bookmaker from the front office came down to the bogfins' dimly-lit cellar (that's old Bell Inc. for you-my candle is getting low and smelly). He said: Pack it in chaps. Don't forget we own half the Varires company and control Pots Inc. You can't invent the noiseless pot. Let's put it to another use, or scrap it.

And so the transistor, as an amplifying device, was reluctantly dragged forth.
Well, that's my story. But the truth was probably quite different. Most inventions are the result of sweated persistence. 'Eureka' was probably Greek for 'the water's too hot.'

Henry has seen some of this dogged persistence in action. A great friend is an industrial de-


You fiddled the meters
signer. His development chief presents him with impossible problems a couple of times a week. He is given a pretty free hand-enviable job!- and the limiting factor is the utmost reliability and speed of construction. No jerry-built hook-ups for him. His tobacco-tin wizadry is made of cast-iron.
So I wonder how he is getting along in the dim-out? I only hope he did not try to follow the same circuit my colleague tried with the cadmium sulphide cell. Published, he told me, in $\mathrm{Pr}^{*} \mathrm{ct}^{*} \mathrm{c}^{*}$ I ${ }^{*} 1^{*} \mathrm{ctr}^{*} \mathrm{n}^{*} \mathrm{cs}$ (R**ly? Ed.). It did not work, and there was no time to find out why.
So this chap being in his way a latter-day Brattain, persisted in racking his brains for past applications and came up with the trigger circuit for an oven control circuit. Problem-no unijunction transistor. Solution, use a pnp in an involuted hookup-good old Shockley's method.
The final result, with two transistors and four resistors, worked a treat. But as the CDS had to be the lower limiting factor of his resistor network, he now has the problem that there is enough infra-red reflected off the ceiling from a television screen to latch the circuit on.
So what, you say, if there is no power, there will be no telly.
That's where you'd be wrong, you see. The first thing he did was to rig up a battery driven invertor to feed the television receiver!

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IN the past a number of utility type i.c.'s were reviewed in the "I.C. of the Month" series such as the Mullard TAA320 Bifet or the D13V voltage regulator, etc., and it is quite some time now since any of these three or four terminal devices have appeared. Consequently this month's article is devoted to a review of two specialised monolithic i.c.'s namely the Silicon Unilateral and Bilateral Switch (SUS and SBS) which have recently been released by the General Electric Company. Some useful practical applications of the devices are given.

## Operation

In appearance the SUS and SBS resemble a typical small signal transistor mounted in an epoxy TO-98 package, the electrical characteristics of the SUS closely resembling those of an ordinary four layer thyristor. The monolithic chip has a pair of complementary transistors, a zener diode and a resistor


Fig. 1: Circuit symbol (a), equivalent electrical circult (b), characteristics (c) and outline (d) of the Unilateral Switch 2N4990.
fabricated into it as is shown in Fig. 1. The resulting static electrical characteristics of the device, in which the applied voltage versus current is plotted, is also illustrated and from this graph it can be seen that once the applied voltage exceeds a typical value of about 8 volts rapid forward conduction takes place with the voltage across the device dropping almost to zero.

## Applications

A gate lead is also provided to allow triggering at lower voltages in addition to providing transientfree wave forms, but in most applications this lead


The 2N4990 used as a motor speed control.


Here the 2N4990 SUS is used to generate a rapia puise from a comparatively slow rising input. Can be used as an over-voltage sensor with the pulse firing a thyristor.


The Silicon Bilateral Switch 2N4991 (SBS) in a mains full wave power control unit.
may be left floating. The SUS therefore is admirably suited for such applications as over-voltage sensing, pulse sharpeners, and SCR triggering. In conjunction with an SCR an efficient and sensitive automatic
over-voltage cut-out sensor can be designed very simply and a number of typical applications of the SUS are given.

Just as the Triac can be considered as consisting of a pair of thyristors connected in inverse parallel so also is the SBS related to its counterpart, the SUS. A schematic diagram of the 2 N 4991 is illustrated in Fig. 2 accompanied by its static electrical characteristics. It fires in both directions so that in opera-


Fig. 2.
(CKG12)
tion it resembles the ST2 Diac which has established itself as one of the most popular triggering devices in Triac power control circuits. However the 2N4991 has tighter specifications producing sharper pulses than the ST2 although it is offered in the same price bracket as the latter.

Both units are available from Jermyn Industries Ltd., Sevenoaks, Kent.

## TELEVISION

## MAY ISSUE

## PAL DECODER

This month we start on the construction of the TELEVISION Colour Receiver and go straight to the heart of things-the PAL-D decoder. This is built on a conveniently sized printed circuit board and uses well tried and tested circuitry. The board can of course be used as a separate module with other equipment if the reader wishes. In building this decoder you will have a real opportunity to learn about colour signal decoding through practical experience.

## RENOVATING THE RENTALS

The Pye/Ekco T418/11U series gets our attention this month.

## WIDEBAND BAND 1 AERIALS

Essential for DX use but not commonly available is the wideband Band 1 aerial. Along with details of practical designs Roger Bunney also presents information on the fundamentals of wideband television aerial design.

DAVID RICHARDSON of West Ealing has been active on the medium waves with his Lafayette HA230 and medium wave loop antenna. He reports Radio Popular La Palmas on 836 kHz ; CKBW Bridgewater, Nova Scotia on 1000 kHz ; WHN New York City 1050 kHz ; CJRP Quebec, broadcasting in French on 1060 kHz ; WBAL Baltimore 1090 kHz ; CBI Sydney, Nova Scotia 1140kHz; CKCV Quebec, in French, 1280 kHz ; WABK Gardiner, Maine also on 1280 kHz ; WLAM Lewiston, Maine 1470 kHz ; WCKY Cincinnati, Ohio 1530 kHz . David has also logged the Voice of America Okinawa 1178 kHz at 1445 hrs SINPO 22432 , with a programme in Chinese. This outlet is usually heard in Europe during the afternoon in winter. It has a power of 1000 kW , it beams westward into China and sometimes competes with Horby, Sweden on the same frequency when conditions are good.
W. Pennington who lives at Grimethorpe near Barnsley in Yorkshire has built the Radio Nederland MW Frame (Loop) aerial. His receiver is a B40D and on Tuesday February 8th he heard CJON St John's on 930 kHz , WINS New York 1010 kHz ; CBA Moncton, New Brunswick 1070 kHz . After midnight he logged CBN 640 kHz in St John's; CJOX 710kHz Grand Bank, Nfid; Radio Demerara Guyana 760 kHz ; WHN 1050 kHz and WNEW 1130 kHz both in New York City. "I thought it was impossible to receive these stations in England but I have proved it for myself and I find it most interesting" he writes and he suggests that readers may like to write to Radio Nederland, Postbus 222, Hilversum, Holland and ask for a copy of The Frame Aerial leaflet, which gives constructional information on a medium wave loop antenna.

Mike Kerry of Golders Green sends an interesting $\log$ of Latin Americans heard on his Eddystone 750 and medium wave loop. These include YVQQ Radio Puerto Cruz, Venezuela on 760 kHz ; Radio Demerara 760 kHz located in Georgetown Guyana, with programme in English; R.Oriental, Uruguay 770 kHz ; La Voz del Rio Cauca, Colombia 820 kHz ; Radio Mundial, Rio de Janeiro 860 kHz ; R.Tamandau, Recife 890 kHz ; Radio Jornal do Brasil 940 kHz ; Radio Belgrano, Buenos Aires 950 kHz ; R.Sutatenza, Colombia 960 kHz ; Radio Record, Sao Paulo 1000 kHz ; R.Margarita, Venezuela 1020 kHz ; R.Revolution Cuba 1060 kHz ; R.Nacional, Sao Paulo 1100 kHz ; WBMJ San Juan, Puerto Rico 1190 kHz ; R.Valera, Venezuela 1230 kHz ; R.Canaima, Ciudad Bolivar, Venezuela 1290 kHz ; R.Nacional, Santiago in the Dominican Republic 1380 kHz . Mike also mentions the VOA Marathon Key, Florida on 1180 kHz which broadcasts in Spanish. Latin Americans become prominent as the year advances and it is worth while listening between 0100 hrs GMT and sunrise for the stations listed above.

Please send reports and information about the medium waves to the author at 132 Segars Lane, Southport, PR8 3JG.

## TRANSISTORS

| 2G301 | 20p | 2N3404 | $32\}$ p | 40310 | 45p | BC212L | 18p | B8X28 | 82tp | NKT281 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 26302 | 20 p | 2N3405 | 45p | 40311 | 85p | BCY30 | 271p | BEX 60 | $821 p$ | NKT401 |  |
| 2 C 303 | 20p | 2N3414 | $28 . \mathrm{p}$ | 40312 | 471 ${ }^{\text {P }}$ | BCY31 | 80 p | B8X 61 | 621p | NKT402 | 0 p |
| 2G306 | 481p | 2N3415 | 221 p | 40314 | 871 P | BCY3？ | 50p | RSX76 | 2210 | NKT403 | 75p |
| 2 G 308 | 80 p | 2N3416 | 8710 | 40320 | 471p | BCY33 | 25p | BSX77 | 27 | NKT404 | 627 |
| ${ }^{2} \mathrm{G3} 309$ | 80 p | 2N3417 | 87 18 | 40323 |  | BCY34 | 80p | BSX 78 | $27 \pm$ | NKT405 | 750 |
| $2 \mathrm{Cl371}$ | 15p | 2N3570 | \＄1．25 | 40324 | 47 p | BCY38 | 40p | B8Y10 | $27 \pm$ | NKT406 | 32tp |
| 2 C 374 | 20p | 2N3572 | 97 p | 40326 | 3710 | BCY39 | 60 p | BSY11 | 275 | NKT451 | 621p |
| ${ }^{2} \mathbf{G 3 8 1}$ | 201 p | 2N3603 | 271p | 40329 | 80p | BCY 40 | 60p | BSY 24 | 15 p | NKT452 | 62tp |
| 2N404 | 2815 | 2N3606 | 8715 | 40344 | 271 p | BCY42 | 15p | BSY25 | 17p | NKT453 |  |
| 2N698 | 20 p | 2N3807 | 221p | 40347 | 571 p | BCY 43 | 15p | BSY 26 | 173 | NKT603F | 323p |
| 2N697 | 17p | 2N3702 | 11 p | 40348 | 52\％p | BCY54 | 82tp | B8Y27 | 17 p | NKT613F | 82才］ |
| 2N698 | 85 p | 2N3703 | 10D | 40360 | 481 p | BCY58 | 22tp | B8128 | 17 pp | NKF674F |  |
| 2 N 706 | 12¢p | 2N3704 | 11 p | 40361 | 47 p | BCY59 | $28 \pm$ | B8Y29 | 179 | NKT677 |  |
| ${ }^{2} \mathrm{~N} 705 \mathrm{~A}$ | 1210 | 2N3705 | 10p | 40362 | $57 \pm$ | BCY60 | 971 p | B8Y32 | $25 p$ | NKT713 | 5 |
| 2N708 | 15 p | 2N3706 | ${ }^{09 p}$ | 40370 | $82 \pm$ | BCY70 | 20 p | B8Y36 | 25p | NKT781 |  |
| 2N709 | 621p | 2N3707 | 11p | 40406 | 57tp | BCY71 | 25 | BSY37 | $25 p$ | NKT1041 | Op |
| 2N718 | 250 | 2N3708 | 078 | 40407 | 40p | BCY72 | 17p | BSY38 | 22tp | NKT1 |  |
| ${ }^{2} \mathrm{~N} 786$ | 809 | ${ }^{2}$ N3709 | $0^{098}$ | 40408 | 59tp | BCZ10 | 2710 | BSY 39 | $22+5$ |  | 18 |
| 2N727 | 80p | 2N3710 | 090 | 40410 | 621 p | BCZ11 | 42. | BSY40 | 824 | NKT105 |  |
| ${ }_{2} \mathrm{Na}_{14}$ | 17 pp | 2N3711 | 12 p | 40467A | 579 | BD116 | 31－124 | BSY51 | 824 p |  | 32 |
| ${ }^{2}$ 2N916 | 179 | 2N3715 | 21．25 | 40468.4 | 85 p | BD12］ | 85， | BgY52 | 3210 | NKT20 |  |
| ${ }^{2} \mathrm{~N} 918$ | ${ }^{80} \mathrm{p}^{\text {p }}$ | 2N3716 | 21.80 | 40600 | 57p | BD123 | 821 p | B8Y53 |  |  | 1p |
| 2N929 | 22 2p | 2N3791 | 82．08 | AC107 | 80 p | BD124 | 60p | BSY54 | 40p | NKT203 |  |
| 2N930 | 27 pp | 2N3819 | 85 p | AC126 | 20p | BD131 | 75 p | BSY56 | 90p |  | 37\％ |
| 2N1090 | 28.5 | 2N3823 | ${ }^{971 p}$ | ${ }^{\mathbf{A} C 127}$ | 250 | BD132 | 85p | BSY78 | 710 | NKT80 |  |
| 2N1091 | $22 \pm$ | 2N3854 | 878 | ACl28 | 200 | BDY 10 | 81.871 | BSY79 | 45p |  | 773p |
| 2N1131 | 250 | 2 N 3854 A | 2719 | AC154 | $22]$ p | BDY11 | 81.62 | BSY82 | $52 \pm$ | NKT801 |  |
| 2 N 1132 | 25 p | 2N3855 | 2715 | AC176 | 25p | BDY17 | 21.50 | RSY90 | 573 |  | 978p |
| 2 N 1302 | 17 p | 2N3855A | ${ }^{80}{ }^{\text {p }}$ | AC187 | 621 p | BDY18 | 21.75 | B8Y95A | 12dp | NKT801 |  |
| 2N1303 | ${ }^{179} 9$ | ${ }^{2} \mathrm{~N} 38858$ | 30p | ${ }_{\text {AC188 }}$ | 878 | BDY19 | ${ }^{\text {c1．}} \cdot 671$ | BSW 41 | 427 p |  | 1－12 |
| ${ }_{2}^{2 N 1304}$ |  | 2N3856A | 85p | ACY17 | 27 p | BDY20 | 81.121 | R8W70 | 275 | NKT8 |  |
| ${ }_{2}^{2 N 1305}$ | 22tp | ${ }^{2} \mathrm{~N} 3858$ | 25 D | ACY18 | 25 p | BDY38 | 971 p | ${ }_{\text {C11 }}$ | 765 |  | 1 |
| ${ }^{2} \mathrm{~N} 1306$ | 250 | 2N3858A | 80D | ACY19 | 85p | BDY 60 | 81.85 | C424 |  | NKT802 |  |
| 2N1307 | 250 | ${ }^{2} \mathrm{~N} 3859$ | 878 | ACY 20 | 25p | BDY61 | 81.85 | ${ }^{\text {C425 }}$ | 55p |  | 021p |
| ${ }^{2} \mathrm{~N} 1308$ | 800 | 2N3859A | 8810 | ACY21 | 25 p | BDY62 | 81.00 | C426 |  | NKT802 |  |
| ${ }^{2} \mathrm{~N} 1309$ | 800 | 2N3860 | 30 p | ACY 22 | 20 p | BF115 | ${ }^{26 p}$ | C428 | 37 tp |  | 02］p |
| 2N1507 | 17 p | ${ }_{2} \mathbf{N} 3866$ | 11.50 | ACY 28 | 20p | BF117 | 477 | C744 | 80 | NKT802 |  |
| ${ }_{2} 2 \mathrm{~N} 1613$ | 250 | 2N3877 | ${ }^{40} \mathrm{D}$ | ACY40 | 20 p | BF163 | 87 p | D16P1 | 877 |  | 219 |
| 2 Nl 1631 | 85 p | 2 N 3877 A | ${ }^{40} \mathrm{p}$ | ACY41 | 25p | BF167 | 18 p | D16P2 | 40p | NKT802 |  |
| ${ }_{2}^{2 N 1632}$ | 30p | ${ }^{2} \mathrm{~N} 3990$ | 875 | ACY44 | 40 p | BF173 | ${ }^{19 p}$ | D1683 | 877 |  | 82］p |
| ${ }_{2}^{2 N 1638}$ | ${ }^{2770}$ | $\begin{aligned} & \text { 2N3900A } \\ & \text { 2N } 3901 \end{aligned}$ | ${ }^{479}$ | AD140 | ${ }^{\text {62，}}$ | BF177 | ${ }^{30}$ p | D16P4 |  | NKT80 |  |
| 2N16711 | 81.00 | 2N 3903 | 8．59 | AD150 | ${ }_{627} 5$ | BF178 BF 179 | p | GET113 | ${ }^{30 p}$ | OC20 |  |
| 2N1711 | 259 | 2N3904 | 85 D | AD161 | 371p | BF180 | 35p | GET114 |  | OC22 | p |
| 2N1889 | 821p | 2N3905 | 877 | AD162 | $87 \pm$ | BF181 | 324 | GET118 |  | 0 O 23 | 60 p |
| 2N1893 | 878 | 2N3096 | 87 p | AF106 | 42］p | BF184 | 25p | GET119 | 20p | OC24 | ， |
| ${ }^{2} \mathrm{~N} 2147$ | 88.5 | 2 N 4058 | 178 | AF114 | $25^{2}$ | BF185 | 421p | GET120 | 52tp | OC25 | p |
| ${ }^{2} \mathrm{~N} 2148$ | ${ }^{571 p}$ | 2 N 4059 | 109 | AF115 | 25p | BF194 | 17p | GET873 | 12 D | OC26 | 271p |
| 2N2160 | $57 \frac{1}{4}$ | 2 N 4060 | 12 t | AF116 | 25p | BF195 | 15p | GET880 | 80 p | OC28 | 62tp |
| 2N2193 | 40p | 2 N 4061 | $121 p$ | AFl 17 | 25p | BF196 | 482 | GET887 | 20 p | OC29 |  |
| 2N2193A | 42pp | 2 N 4062 | 12\％p | AF118 | 624 | BF197 | 42tp | GET889 | $22^{2}$ | OC35 | p |
| ${ }_{2 N}{ }^{2} 2194$ A | 309 | 2 N 4244 | $471 p$ | ${ }^{\text {AF119 }}$ | 80 p | BF198 | 421 p | GET890 | 22 | OC36 | 62 yp |
| ${ }^{2 N} 2217$ | 275 | ${ }^{2} \mathrm{~N} 4285$ | 175 | AF124 | $22 . p$ | BF200 |  | GET896 | 22\％p | OC41 | $221 p$ |
| ${ }_{2}^{2 N} 22218$ | ${ }^{238}$ | ${ }_{2}^{2 N} 4286$ | 17p |  | ${ }^{20}$ | BF224 | 14 p | GET897 | $22 t$ | ${ }_{0} \mathrm{OC42}$ | ${ }^{20}$ |
| 2 N 2219 | ${ }^{235}$ | 2 N 4287 | 171 | AF126 | ${ }^{200}$ | ${ }^{\text {BF225 }}$ | 19p | GET898 | 224 | OC44 | 0p |
| ${ }_{2} \mathbf{N} 2220$ | ${ }^{25}$ | 2 N 4288 | 17tp | AF127 | 17p | BF237 | 28p | MJ400 | $81.07 \frac{1}{2}$ | $0 \mathrm{OC4}$ | 12tD |
| 2 N 2221 | 250 | 2 N 4289 | ${ }^{17}{ }^{17}$ | AF138 | 872p | BF238 | 28p | MJ420 | 21－12t | $0 \mathrm{OC4}$ | 15p |
| ${ }_{2}^{2} \mathrm{~N}_{2} \mathrm{~N}_{2970}$ | ${ }^{30 \mathrm{p}}$ | ${ }_{2}{ }^{2 N} 4290$ | 1718 | AF178 | 48.1 | ${ }^{\text {BFP244 }}$ | 88p | MJ421 | E1－12\％ | OC70 | 159 |
| ${ }^{2} \mathbf{2 N 2 9 7 0}$ | 478 30 p 3 | ${ }_{2}^{2 N 4291}$ | ${ }_{18}^{178}$ | AF179 AF180 | 7248 58 | $\mathrm{BFW61}_{\text {BFX }}$ | 472 p 2280 | MJ439 | 21．021 | OC71 | 1218 |
| ${ }^{2 N} 2368$ | 17 p | 2N4303 | 478 | AF181 | 484 | BFX13 | 22pp | MJ480 | ${ }_{97 \mathrm{tp}}^{95}$ | ${ }_{0} \mathrm{C} 74$ |  |
| 2 N 2369 | 17p | 2N5027 | 6215 | AF239 | $42+\mathrm{p}$ | BFX29 | 30 p | M．5481 | 红1．25 | OC75 |  |
| ${ }^{2}{ }^{2} 2369 \mathrm{~A}$ | 17p | 2N5028 | 578 | AF279 | 47 p | BFX30 | 80 p | MJ490 | 11．00 | OC76 | $22+p$ |
| 2N2410 | 42.5 | 2N5039 | 479 | AF280 | 624 p | EFX42 | 37 p | MJ491 | 81.371 | ${ }^{0} \mathrm{C} 77$ |  |
| 2 N 2483 |  | ${ }^{2 N 5030}$ | 481p | AF＇211 | 324 | BFX44 | 871p | MJ1800 | 22－17 | OC81 | 20 p |
| ${ }_{2} 2 \mathrm{~N} 2484$ | 8910 | ${ }_{2} \mathbf{2 N 5 1 7 2}$ | 12.1 | ABY26 | 85 p | BFX68 | ${ }^{67 p}$ | MJF340 | 6218 | OC81D | $22 \pm$ |
| 2N2639 | ${ }_{20}^{28}$ | 2N5174 |  | ASY27 | $87{ }^{87}$ | BFX 84 | 25 p | MJE520 | 60p | $0 \mathrm{OC83}$ | 25 |
| 2N2540 | 28. | ${ }_{2}^{2 N 5175}$ | 5215 | AEY28 | 2718 | ${ }_{\text {BFX }}{ }^{\text {BF }}$ | 824 | MJE621 | $73 p$ | ${ }^{0} \mathrm{C} 84$ | 25 p |
| $2 \mathrm{~N} 2 \mathrm{Al3}$ | 85 p |  | ${ }^{45}$ | ASY29 | 27tp | BFX86 |  | MPF102 | 42p | OC139 | $82 \pm$ |
| ${ }^{2 N} 2614$ | ${ }_{50 \mathrm{p}}^{30}$ | 2N5232A | 30 p | ASY36 | 250 | BFX87 | 2715 | MPF103 | 87p | ${ }^{0} \mathrm{Cl} 140$ | 821 p |
| 2N2646 | ${ }^{524}$ | 2N5245 | 45 p | ASY50 | 25p | BFX 88 | 25 p | MPF104 | 377 | ${ }_{0} \mathrm{OC170}$ | 30 p |
| ${ }^{2} \mathbf{2 N 2 6 9 6}$ | ${ }^{82}$ | 2N5246 | ${ }^{421}{ }^{\text {d }}$ | ASY51 | $82 \%$ p | BFX89 | 62.5 | MPF105 | ${ }^{371 p}$ | ${ }^{\text {OC171 }}$ | 300 |
| ${ }_{\text {2N }} \mathbf{2} \mathbf{N} 2712$ | ${ }^{25}$ | ${ }^{2} \mathrm{~N} 5249$ | ${ }^{671}$ | ASY54 | 25p | BFX93A | 70p | MPP3638 | 832 | ${ }^{\circ} \mathrm{CO} 200$ | 40 p |
| 2N2713 |  | ${ }_{2}{ }_{2} \mathrm{~N} 52656$ | ${ }_{82}^{28.75}$ | AsY86 | ${ }^{82}+$ | BFY10 | 32 tp | NKT0013 | 3 471p | ${ }^{\text {OC201 }}$ | ${ }^{60} \mathrm{D}$ |
| 2N2714 | 800 | 2N5267 | 82．624 | ${ }_{\text {A }}$ | 42tp | ${ }_{\text {BFY }}{ }^{\text {BFI7 }}$ | 429p 224 p | NKT124 | 42ip |  | ${ }^{75 p}$ |
| 2N286s | 627 D | 2N5305 | 871 D | BC107 | 10 p | BFY18 | $32 \pm$ | NKT126 | 277p | ${ }_{\mathrm{OC} 204}$ | 48 4 4p |
| 2 N 2904 | 809 | 2N5306 | 400 | BC108 | 10p | BFY19 | $32+p$ | NKT128 | 27tp | OC205 | 90 p |
| ${ }^{2} \mathrm{~N} 2904 \mathrm{~A}$ | 82p | 2 N 5307 | $87{ }^{\text {p }}$ | BC109 | 10p | BFY20 | \＆1．60 | NKT135 | 87\％p | OC207 | 750 |
| ${ }_{2} 2 \mathrm{~N} 2905$ | 37 p | 2N5308 | $87{ }^{87}$ | ${ }^{\mathrm{BC} C 11}$ | 15p | BFY24 | 42 p | NKT137 | 32tp | OCP71 | $48 . p$ |
| 2N2905A | ${ }^{40 p}$ | ${ }_{2}^{2 N 5309}$ | ${ }_{4818} 48$ | ${ }_{\text {BC116 }}^{\text {BC15 }}$ | ${ }_{15 p}^{150}$ | BFY24 | 450 | NKTT210 | 800 | ORP12 | ${ }^{50 \mathrm{p}}$ |
| ${ }_{2}^{2 N 29060}$ | 27\％ | ${ }_{2}^{2 N 5310}$ | 427p | ${ }_{\text {BCl18 }}$ | $15 p$ $10 p$ | BFY25 | ${ }^{250}$ | NKT211 | 300 80 p | ORPB4 ${ }^{\text {Pr }}$ |  |
| 2N2907 | 80 p | 2N5355 | 2710 | － $\mathrm{Cl}^{2} 1$ | 20 p | BFY29 | 50 p | NKT213 | ${ }^{30}$ | TIS34 | 22ip |
| 2 N 2923 | 15p | 2N5356 | 82tp | $\mathrm{BCl}^{\text {ch2 }}$ | 20p | BFY 30 | 50 p | NKT214 | 224 p | TIS43 | 27 D |
| 2 N 2924 | ${ }_{15 p}$ | 2N5365 | 478 | ${ }_{\text {BCl2 }}$ | 20 p | BFY41 | 50p | NKT215 | 28 | TIS44 | 100 |
| 2N2925 | 150 | 2N5366 | 321 p | BCl26 | 20p | BFY43 | 6218 | NKT216 | 87p | TIS45 | 10 p |
| 2N2926 |  | 2N5367 | 575 | ${ }^{\mathrm{BC}} 140$ | 3710 | BFY50 | 285 | NKT217 | $48 . p$ | TIS46 | ${ }_{11} 1$ |
| Green | 14p | 2N5457 | 377 | ${ }^{\text {BC147 }}$ | 10 p | BFY51 | 20p | NKT219 | 30 | TIS47 | $11 p$ |
| Yellow | $12+1$ | 28005 | $7{ }^{75}$ | ${ }^{\mathrm{BCl} 48}$ | 10 p | BFY52 | 23 p | NKT223 | 27 p | TTS48 | 18 p |
| Orange | 12 p | 28020 28102 | ${ }_{50.00}^{500}$ | ${ }_{\text {BC152 }}$ | 17p | BFY553 | 17 p | NKT224 | 25 D | TTS49 | 18 p |
| ${ }_{2}^{2 N 3011}$ | 80 p | ${ }_{28102}^{28103}$ | ${ }^{50 \mathrm{p}}$ | ${ }^{\mathrm{BC}} \mathrm{BC} 52$ | 17 p | RFY56A | 575 | NKT225 | 2elp | TIS50 | 177 |
| 2N3014 2 N 3053 | $82+$ | ${ }_{2}^{2 S 103}$ | 25p | ${ }^{\text {BC157 }}$ | 200 | BFY75 | 30p | NKT229 | 80p | TIS51 | 124 p |
| 2N3053 | 18p | 28104 | ${ }^{265}$ | BC158 | 11p | BFY76 | 48.5 | NKT237 | ${ }^{85}$ | TI852 | 121p |
| 2N3054 | ${ }^{\text {48p }}$ | 28501 28502 |  | ${ }_{8 C 159}$ | 120 | BFY77 | ${ }^{5778}$ | NKT238 | 25p | TTS53 | $29 . \mathrm{p}$ |
| 2N 3055 2 N 3133 | 82p | 28502 28503 | ${ }_{27+\mathrm{p}}^{85 \mathrm{p}}$ | ${ }_{\text {BC167 }}^{\text {BC1 }}$ | 62ip | BFY\％90 | ${ }^{67 \%}$ | NKT240 | 87 p | T1880 | $22+p$ |
| 2N3134 | 30 p | 3N83 | ${ }_{40}$ | BC168B | 10 p | BFW59 | 275 | NKT241 | ${ }^{271 p}$ | TIS61 | ${ }_{27 \text { cp }}^{26 p}$ |
| 2N3135 | 25p | 3N128 | 70p | BC168C | 11 p | ． $\mathrm{BFW}^{\text {6 }} 0$ | 250 | NKT243 | 621 | TIP29A | p |
| 2N3136 | 25 p | 3N140 | 779 | BC1698 | 11 p | － $\mathrm{BPX}^{25}$ | 21.85 | NKT244 | 17p | TIP30A |  |
| 2N3390 | ${ }^{26 p}$ | 3N141 | $72 ¢$ | ${ }^{\text {BC169C }}$ | 12p | BPX29 | 21.80 | NKT245 | 200 | Tip31a | 62 ${ }^{\text {p }}$ |
| 2N3391 | 20 p | ${ }^{3 N 142}$ | 65p | BC170 | 12tp | BPY10 | E1．45 | NKT261 | 20p | TIP32A | 759 |
| ${ }^{2} \mathrm{~N} 3391 \mathrm{~A}$ | 80p | 3N143 | 67 p | BC171 | ${ }^{15 p}$ | BRY39 | 87 p | NKT262 | 20 p | TIP93A |  |
| 2N3392 | 17p | ${ }^{3 N 152}$ | 87 \％p | ${ }^{\text {BC172 }}$ | 15p | BSX 19 | 1710 | NKT264 | $20 p$ |  | 1.02 p |
| ${ }_{2}^{2} \times 3.5393$ | ${ }^{15 p}$ | R．C．A． | 58.7 | ${ }^{\text {BC175 }}$ | 22 p | BSX20 | 1718 | NKT271 | 20 p | TIP34A | 42.05 |
| 2N3394 | ${ }_{22+p}^{15 p}$ | 40050 40251 | ${ }_{82+5}$ | ${ }_{\text {BC183 }}$ | ${ }^{10 p}$ | BSX21 | 8715 | NKT272 | 20 | T1P35A | 22.90 |
| 2N3402 | 22ip | 40251 | 3275 | BC183 | 09p | B8x26 | 45p | NKT274 | 20p | TIP36A | 23－68 |
| 2N3403 | 224 P | 40309 | 32łp | BC184 | 11p | B8X27 | 4710 | NKT275 | 20 p |  |  |
|  | Post \＆Packing 13p per order．Europe 25p．Commonwealth（Air）65p（MIN．） <br> Matching charge（audio transistors only）15p extra per pair． Prices subject to alteration without prior notice． |  |  |  |  |  |  |  |  |  |  |

TTL．LOGIC I．C．NEW PRICES

|  | $\begin{array}{r} 11 \\ \text { en } \end{array}$ | $\begin{gathered} 12-24 \\ 8 p \end{gathered}$ |  | $\begin{array}{r} 1-11 \\ 8 p \end{array}$ | $\begin{gathered} 12-24 \\ 8 \mathrm{p} \end{gathered}$ |  | $\begin{array}{r} 1-11 \\ 8 \end{array}$ | $\begin{array}{r} 12-24 \\ 5 \mathrm{p} \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SN7400 | 0.20 | 0.18 | EN7438 | 0.80 | 0.75 | SN 7472 | 0.88 | 0.80 |
| 8N7401 | 0.80 | 0.18 | SN7437 | 0.64 | 0.06 | SN7473 | 0.48 | 0.41 |
| 8N7402 | 0.20 | 0.18 | 日N7438 | 0.64 | 0.60 | SN7474 | 0.48 | 0.41 |
| 8N7403 | 0.80 | 0.18 | SN7440 | 0.28 | 0.81 | SN7475 | 0.45 | 0.44 |
| SN7405 | 0.80 | 0.18 | SN7441AN | 0.87 | 0.88 | SN7476 | 0.45 | 0.44 |
| 8N7406 | 0.80 | 0.75 | SN7442 | 0.85 | 0.81 | GN7480 | 0.70 | 0.65 |
| 8N7407 | 0.80 | 0.75 | SN7443 | 2.88 | 8.70 | SN7481 | 1.40 | 1.88 |
| SN 7408 | 0.20 | 0.18 | SN 7444 | $8 \cdot 86$ | $8 \cdot 70$ | SN7482 | 0.87 | 0.88 |
| SN7409 | 0.20 | 0.18 | SN7445 | 8.50 | 8.40 | GN7483 | 0.87 | 0.89 |
| 8N7410 | 0.20 | 0.18 | SN7446 | 1.00 | 0.98 | 8N7484 | 8.00 | 1.85 |
| 8N741t | 0.88 | $0 \cdot 21$ | EN7447 | 1.00 | 0.85 | SN7485 | 8.68 | 8.40 |
| SN7412 | 0.48 | 0.48 | SN7448 | 1.00 | 0.95 | SN7486 | 0.38 | 0.80 |
| SN7413 | 0.40 | 0.88 | 8N7449 | 1.00 | 0.95 | SN7490 | 0.87 | 0.84 |
| 8N7420 | 0.20 | 0.18 | GN7450 | 0.20 | 0.18 | SN7491AN | $1 \cdot 21$ | 1.10 |
| SN7423 | 0.51 | 0.47 | SN7461 | 0.20 | 0.18 | SN7492 | 0.87 | 0.84 |
| SN 7427 | 0.48 | 0.45 | SN7453 | 0.20 | 0.18 | 8N7493 | 0.87 | 0.81 |
| 8N7428 | 0.80 | 0.75 | gN7454 | 0.20 | 0.18 | SN 7494 | 0.87 | $0 \cdot 84$ |
| SN7430 | 0.28 | 0.15 | EN7460 | 0.20 | 0.18 | 8N7495 | 0.87 | 0.84 |
| 8N7432 | 0.48 | 0.42 | SN7470 | 0.40 | 0.88 | SN7496 | 0.87 | 0.81 |

MULLARD SUB－MIN ELECTROLYTIC
 $6.4 / 6 \cdot 4 ; 6.4 / 25 ; 8 / 40 ; 10 / 16 ; 10 / 64 ; 12 \cdot 5 / 25 ; 16 / 40 ; 20 / 16 ; 20 / 64 ; 25 / 6 \cdot 4 ;$ $25 / 25 ; 32 / 10 ; 32 / 40 ; 32 / 64 ; 40 / 16$ ； $60 / 6.4$ ； $50 / 25$ ； $50 / 40$ ；64／10；80／2．5； $80 / 16 ; 80 / 25 ; 100 / 6 \cdot 4 ; 125 / 10$ ； $125 / 16$ ； $200 / 10$.

## SILICON RECTIFIERS

| PIV | 50 | 100 | 200 | 400 | 000 | 800 | 1000 | 1200 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1A | 8p | 日p | $10 p$ | 11p | 19p | 15p | $80 p$ | － |
| 3A | 15p | 17p | 20 p | 281 p | $25 \%$ | 27p | 80 p | $85 p$ |
| 6A |  |  | 26p | 30p | 82 tb | 85 |  |  |
| 10A | － | $681 p$ | 57 ip | 65p | 7\％1p | 861 p | 9710 | 21.85 |
| 15A | － | 578 p | 62tp | 775 | 909 | 9715 | 81.80 | 81．671 |
| 35A |  | 800 | 90p | 81.00 | \＄1．26 | 81.50 | 28.50 | －－ |
| 1 amp and 3 amp are plastic encapsulation． |  |  |  |  |  |  |  |  |

DIODES \＆RECTIFIERS

| IN34A | 10 p | AA119 | 7 p | BAX16 | 121p | FST3／4 | 28\％ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IN914 | 7 D | AA129 | 15p | BAY18 | 17tp | OA5 | 17p |
| ${ }^{\text {IN }} 916$ | 7 p | AAZ13 | 12p | BAY31 | 70 | OA10 | 20p |
| IN 4007 | 20p | AAZ15 | 12． | BAY38 | 25p | OA9 | 10 p |
| IS44 | 7p | AAZ17 | 10p | BY100 | 15p | OA47 | 8 p |
| 18113 | 15p | BA100 | 15p | BY103 | 22p | OA70 | 7 p |
| 18120 | 12p | BA102 | 25 p | BY122 | 47 p | OA73 | 10p |
| 18121 | 14p | BA110 | 25p | BY124 | 15p | OA79 | 7 |
| 18130 | 8 p | BA114 | 15p | BY126 | 15p | OA81 | 8 |
| 18131 | 10 p | BA115 | 7 p | BY127 | 17p | OA82 | 10 p |
| 18132 | 12D | BA14］ | 17p | BY164 | 57p | OA99 | 7p |
| 18920 | 7 D | BA142 | 17p | BYX10 | 22p | OA9］ | 7 |
| 18922 | 8 D | BA144 | 120 | BYZ10 | 35p | OA95 | 7 |
| 18923 | 12p | BA145 | 17p | BYZ11 | 38p | OA200 | 7 |
| 18940 | 5 p | BA154 | $\underset{5 p}{18 p}$ | BYZ12 | 300 800 | OA202 TIV307 | 10 p 50 |

## OPTOELECTRONICS

MINITRON 3015F 7－SEGMENT INDICATOR（ 16 PIN DIL）E2．00 GNP•7AH COLD CATHODH TVBE SIDE VIEWING． $0-9$ and TWO DECIMAL POINTS．May be
driven by SN744／AN．75p

TIL 209 LIGHT EMITTING OIODF．Made by TEXAS INST． （Red）．85p
B9900 PEOTORESISTOR 88p

## VEROBOARD

|  | $\begin{aligned} & 0.15 \\ & \text { Matrix } \end{aligned}$ | $\underset{\text { Matrix }}{0.1}$ |
| :---: | :---: | :---: |
| $21 \times 3$ \％in | 17p | 88 p |
| 2）$\times 5$ in | 25p | 25p |
| $3{ }^{2} \times 3$ in | 85 | 85 |
| $34 \times 5$ in | 800 | 29p |

$5 \times 17$ in（Plain）89p
Vero Pins（Bag of 36 ） 80 p
Vero Cutter 450
Pin insertion T
＂SCORPIO＂CAP
DISCHARGE IGNITION
SYSTEM
（As published in P．E．Nov．
＇7I）．Complete kit £IO．00
P．P．50p．

RESISTORS

BRIDGE RECTIFIERS

| A．PIV |  | A．PIV |  |  |
| :---: | :---: | :---: | :---: | :---: |
| 100 | 87p | 4 | 50 | 60 |
| 1.4140 | 57p | 4 | 100 | 70p |
| 50 | 88p | 4 | 400 50 | 809 |
| 200 | 41p | 6 | 200 | 800 |
| 400 | 46p | － | 400 | E1．10 |

## MULLARD C2

$0.01,0.022,0.083,0.047$ 8p each
$\begin{array}{llll}0.01, & 0.022, & 0.033, & 0.047 \\ 0.068, & 0.10 & \cdots & \cdots\end{array}$ $0.15,0.22,0.33$
0.68
$1 \mu \mathrm{~F}$
$1.5 \mu \mathrm{~F}$
WIRE－WOUND RHSIETORS
2.5 watt $5 \%$（up to 270 ohms

5 watt $5 \%$（up to $8 \cdot 2 \mathrm{k} \Omega$ only）， $9 p$
10 watt $5 \%$（up to $28 k \Omega$ only）．

## POTENTIOMETERS

Carbon：
Lag．or Lin．，less awitch， 169
Log．or Lin．，with switch， $8:$
WIre－wound Pots（ $\mathbf{3 W}$ ）， 88 p
Wire－wound Pots（3W），38p
Twin Ganged Stereo Pots，Log．
PRESETS（CARBON）
${ }_{0}^{0.1}$ Watt $\begin{array}{lll}\text { Bp } & \text { VERTICAL }\end{array}$


## THERMISTORS

$\begin{array}{lll}\text { R63（STC）} & \text { s1．871 } & \text { VA3705 } 95 p \\ \text { K161（1k）} & 189 & \text { VA1077 } 80 p\end{array}$
 stock．Please enquire．


AMPEX 7.5v. D.C. MOTOR. This is an ultra-precision tape motor designed for use in the AMPEX model AG20 portable recorder. Torque $450 \mathrm{GM} / \mathrm{CM}$. Scall load at, 500 ma . Draws 60 ma on run. $600 \mathrm{rpm}+5 \mathrm{a}^{4}$ speed adjustment, internal AF/RF suppression $t^{\prime \prime}$ dia. $I^{\prime \prime}$ spindle, motor $\mathbf{3}^{\prime \prime}$ dia. It". Original cost £16.50. Our price \&4.25. P. \& P. 25p. Mu-metal enclosure available 75p each.

SYNCHRONOUS MOTORS
SMITHS. 250v. 50 Hz . Available in the following R.P.M. 2-3-6-10-30-60. Price 75p each. Carr. paid. CROUZET. 220/380v. 50/60 Hz. 250-300 rpm.
73p. Carr. paid. f't $^{\frac{1}{2}}$ spindle, weight $\&$ lb. Powerful. 88p each.
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IT was on Wednesday, April 10th, sixty years ago that the s.s. Titanic left Southampton on her maiden voyage. Five days later, she sank to the bed of the ocean-a useless mass of twisted and torn steel, for on Sunday, April 14th, 1912, at about 11.40 p.m. the greatest peacetime disaster in maritime history took place when this ship, gross tonnage 46,328 , struck an iceberg, $50-100 \mathrm{ft}$. high, while on her maiden voyage to New York.

The blow on the starboard side of the ship ripped a hole 300 feet long and made many of the essential watertight compartments useless. She foundered within three hours, in water two miles deep.

Mr. H. S. Bride, the operator who survived the disaster described the way in which wireless played its all important role in saving those who were rescued. Mr. Bride was the second operator on the Titanic and was relieving his chief, Mr. Phillips at the time when the collision occurred. "I was standing by Mr. Phillips telling him to go to bed when


Jack G. Phillips, Chief Marconi Operator on board the Titanic.

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How the maiden voyage of the Titanic was advertised.
the captain put his head into the cabin. 'We have struck an iceberg!' he said, 'you had better get ready to send out a call for assistance. 'Don't send it until I tell you'. The captain went away but in ten minutes came back. We could hear terrible confusion outside but not the least thing to indicate any trouble. The wireless was working perfectly. 'Send a call for assistance' ordered the captain. 'What call shall I send?' Phillips asked. 'The regulation international call for help-just that' was the reply, and Phillips began to send the signal C.Q.D. (now replaced by S.O.S.). After a few minutes the Captain reappeared and said, 'Send S.O.S.-it may be your last chance'."

The Titanic's C.Q.D.'s and S.O.S.'s were first picked up by the German steamer, Frankfurt which was 153 miles away. Almost at the same time, the Carpathia's wireless operator reported the emergency to his captain and was able to give the stricken liner's position 4146 N 5014 W . Immediately, the Carpathia, which was 58 miles away, altered course to the rescue. The first message the Carpathia picked up was, "COME AT ONCE, WE'VE STRUCK A BERG. IT'S C.Q.D. OM C.Q.D." C.Q.D. of course meant Come Quick, Danger, and was rather tricky to tap out in Morse. For this reason, the now universal distress signal S.O.S. was also used and this was the first occasion it had really been broadcast in earnest.

Fortunately, the Carpathia's only wireless operator, although officially off watch had returned to
his equipment and was putting out some routine traffic calls-including some to the Titanic. At 12.20 a.m. the distress calls were received and soon at least six ships were steaming to the disaster zone.
At 2.20 a.m. the Californian which for sometime past had been watching the lights of the 'unknown' vessel grow steadily dimmer, noted that they had finally disappeared. The optical illusion being that of seeing a ship disappearing into the darknessnot going under the waves. Those witnessing the scene from the decks of the Californian were not to know that they had been watching the death throes of the Titanic and that the sea around them was dotted with hundreds of human beings fighting for their lives. Two long hours were to elapse before the Carpathia arrived and began to pick up survivors. Among over 1,500 people lost on that fateful night was Jack Phillips, the Titanic's senior wireless operator, who remained at his post as the decks were awash. Mr. Bride, who showed an equal devotion to duty, was eventually rescued after nearly two hours in the sea.

As the decks were awash, Phillips was standing in the wireless room, still sending away giving the Carpathia details of how the Titanic was faring. He then picked up the Olympic and told her that Titanic was sinking by the head. As Phillips was sending the message, Harold Bride strapped his lifebelt to the wireless operator's back. He had already put on his overcoat and was wondering if he could get Phillips' boots on him while he was still sending.

The Captain's voice boomed, "Men, you have done your duty. You can do no more. Abandon your cabins now. It's every man for himself. You look out for yourselves. I release you-that's the way of it at this kind of time, every man for himself."

The boat deck was awash. Phillips clung on sending ... sending. He clung for about fifteen minutes after the Captain released him-water washing the floor of the wireless cabin.

H. S. Bride, 2nd Operator on board the Titanic.


Cyril Evans, Marconi Operator on board the Californian.

From aft came the strains of music-the ship's band playing ragtime. Phillips ran aft and made his way to the deck. He swam to a life-raft but lay there exhausted until his last breath failed. Harold Bride later said of Phillips. "He was a brave man and stuck to his key until the very end. If he had had a chance to go to his room and get warmer clothing, as I did, he would probably be alive today. But duty was first with him."

When water flooded the upper decks, Second Officer Lightoller was up to his ankles in water and Captain Smith-a calm grave figure of a man realising that the end could not now be far off-stood calmly on the bridge. He spoke two words to the people who were crowding forward. Words which were a call to our traditions of race and manhood, "Be British!"

Besides calling the Carpathia and Baltic, Phillips had also contacted the Frankfurt again and vessels had also been sent to the rescue by the wireless station at Cape Race. They were however all very far away and even though they made all speed to the stricken Titanic, not one of them was near when the last terrible moment came.

The band still played on. They played until they were waist deep in the cold dark water and in that awful moment, W. Hartley, the conductor spoke to his colleagues and there rang out into the darkness the hymn NEARER MY GOD TO THEE.

Part 2 of the Titanic story will be published in "Going Back" next month.


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Project 60 modules are more versatile - using them you can have anything from a simple record player or car radio amplifier to a sophisticated and powerful stereo tuner-amplifier. Either power amplifier can be used in a wide variety of applications as well as high fidelity. The Stereo 60 pre-amplifier control unit may also be used with any other power amplifier system, as can the AFU filter unit. The stereo FM tuner operates on the unique phase lock loop principle to provide the best ever standards of sensitivity and audio quality. Project 60 modules are very easily connected together by following the 48 page manual supplied free with all Project 60 equipment. The modules are great space savers too and are sold individually boxed in distinctive white and black cartons. With all these wonderful advantages, there remains the most attractive of all - price. When you choose Project 60 you know you are going to get the best high fidelity in the world, yet thanks to Sinclair's vast manufacturing resources (the largest in Europe) prices are fantastically low and everything you buy is covered by the famous Sinclair guarantee of reliability and satisfaction.
Typical Project 60 applications

| System | The Units to use | together with | Cost of Units |
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| Simple battery record player | 2.30 | Crystal P.U., 12 V battery volume control | £4.48 |
| Mains powered record player | Z.30, PZ.5 | Crystal or ceramıc P.U. volume control etc. | £9.45 |
| $20+20 \mathrm{~W}$. stereo amplifier for most needs | $\begin{aligned} & 2 \times 2.30 s, \text { Stereo 60, } \\ & \text { PZ.5 } \end{aligned}$ | Crystal, ceramic or mag. P.U.. F.M. Tuner, etc. | $E 23.90$ |
| $20+20$ W. stereo amplifier with high performance spkrs. | $\begin{aligned} & 2 \times 2.30 s, \text { Stereo } 60, \\ & \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U.. F.M. Tuner. Tape Deck, etc. | £26.90 |
| $40+40$ W. R.M.S. <br> de-fuxe stereo amplifier | $2 \times 2.50 \mathrm{~s}$, Stereo 60 PZ.8, mains trsfrmr | As above | £34.88 |
| Indoor P.A. | Z.50, PZ.8, mains transformer | Mic., guitar, speakers, etc., controls | £19.43 |

[^7]

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now. Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.
SPECIFICATIONS-Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz . Capture retio : 1.5 dB . Sensitivity: $7 \mu \mathrm{~V}$ for lock-in over full deviation. 8quelch level: $20 \mu \mathrm{~V}$. Signal to noise ratio: $>65 d \mathrm{~B}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}$ ( $\pm 1 \mathrm{~dB}$ ). Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation. Stereo decoder operating leval: $2 \mu \mathrm{~V}$. Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S. Oparating voltage: 25-30VDC.
Indicators: Stereo on ;tuning. Size: $93 \times 40 \times 207 \mathrm{~mm}$

## Stereo 60 Pre-amp/control unit



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout. achieving a really high signal-to-noise ratio and excelient tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.
SPECIFICATIONS-Input eansitivitios: Radio - up to 3 mV . Mag. p.u. 3 mV : correct to R.I.A.A curve $\pm 1 \mathrm{~dB}: 20$ to $25,000 \mathrm{~Hz}$. Ceramic p.u. - up to 3 mV : Aux - up to 3 mV . Output: 250 mV . Signel to noise ratio: better than 70 dB . Channel matching: within 1 dB . Tone controls: TREBLE +12 to -12 dB at 10 KHz : BASS +12 to -12 dB at 100 Hz . Front panel : brushed aluminium with black knobs and controls. 8 ize: $68 \times 40 \times 207 \mathrm{~mm}$.

## A.F.U. High \& Low Pass Filter Unit



For use between Stereo 60 unit and two Z .30 s or $\mathrm{Z.50}$ s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid $: 12 \mathrm{~dB} / o c t a v e)$, there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two fitter stages - rumble (high pass) and scratch (low pass). Supply voltage - 15 to 35 V . Current 3 mA . H.F. cut-off ( -3 dB ) variable from 28 KHz to 5 KHz . L.F. cut-off ( -3 dB ) variable from 25 Hz to 100 Hz . Distortion at 1 KHz ( 35 V. supply) $0.02 \%$ at rated output. Size: $66 \times 40 \times 90 \mathrm{~mm}$.


The $Z .30$ and $Z .50$ are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at $15 \mathrm{w}(8 \Omega)$ and all lower outputs. Whether you 8PECIFICATIONS (2.50 un/ts are interchangeab/ Power Outputs
Z.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts.
Z.50 40 watts R.M.S. into 3 ohms using 40 volts:

30 watts R.M.S. into 8 ohms using 50 volts.
Frequency response: 30 to $300,000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$.
use $Z .30$ or $Z .50$ amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.

## with 2.30 s /n all app/ications).

Distortion: $0.02 \%$ into 80 hms .
Signal to nolse ratio: better than 70 dB unweighted. Input sansitivity: 250 mV into 100 Kohms (for 15 W into 8 8 )
For speakers from 3 to 15 ohms impedance.
Size: $14 \times 80 \times 57 \mathrm{~mm}$.

## Power Supply Units



Designed special for use with the Project 60 system of your choice. Use PZ.5 for normal 2.30 assemblies and PZ. 6 where a stabilised supply is essential.

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| 2N708 | 0.10 | AC154 | 0.16 | BF167 | 0.18 |
| 2N708 | 0.15 | AC157 | 0.20 | BF173 | 0.20 |
| 2N753 | 0.85 | AC169 | 0.10 | BF179 | 0.80 |
| 2N929 | 0.88 | AC176 | 0.25 | BF180 | 0.85 |
| 2N930 | 0.25 | AC187 | 0.30 | BF181 | 0.25 |
| 2N997 | 0.80 | AC188 | 0.80 | BF184 | 0.25 |
| 2N1131 | 0.25 | ACY17 | 0.875 | BF185 | 0.20 |
| 2N1132 | 0.85 | ACY18 | 0.20 | BF194 | 0.15 |
| 2N1184 | 1.25 | ACY19 | 0.25 | BF195 | 0.15 |
| 2N1301 | 0.40 | ACY20 | 0.20 | BF196 | 0.20 |
| 2N1302 | 0.75 | ACY21 | 0.80 | BF197 | 0.20 |
| 2 N 1304 | 0.25 | AGY22 | 0.15 | BF200 | 0.86 |
| 2N1305 | 0.25 | AD140 | 0.60 | BFW87 | 0.25 |
| 2N1306 | 0.25 | AD149 | 0.50 | BFW 88 | 0.88 |
| 2N1307 | 0.80 | AD161 | 0.85 | BFW89 | 0.80 |
| 2N 1308 | 0.85 | AD162 | 0.35 | BFW91 | 0.80 |
| 2N1309 | 0.80 | ADZ11 | 1.25 | BFX88 | 0.25 |
| 2N1613 | 0.22 | ADZ12 | 1.25 | BFY17 | 0.40 |
| 2N1711 | 0.25 | AF114 | 0.20 | BFY19 | 0.60 |
| 2N1756 | 0.75 | AF115 | 0.20 | BFY50 | 0.85 |
| 2N2147 | 0.75 | AF116 | 0.20 | BFY51 | 0.20 |
| 2N2160 | 0.65 | AF117 | 0.20 | BFY52 | 0.25 |
| 2N2217 | 0.80 | AF118 | 0.45 | BgY26 | 0.80 |
| 2N2218 | 0.80 | AF125 | 0.25 | BSY27 | 0.20 |
| 2N2219 | 0.85 | AF127 | 0.20 | B8Y28 | 0.20 |
| 2N 2369 A | 0.20 | AF180 | 0.85 | BSY65 | 0.20 |
| 2N2477 | 0.65 | AF'181 | 0.85 | B8Y95A | 0.16 |
| 2N2646 | 0.60 | AF186 | 0.40 | OC16 | 0.60 |
| 2N2905 | 0.85 | AF639 | 0.40 | OC22 | 0.60 |
| 2N2923 | 0.15 | AFZ11 | 0.45 | OC23 | 0.60 |
| 2N2924 | 0.15 | A8Y28 | 0.25 | OC24 | 0.60 |
| 2N2926 | 0.125 | ASY27 | 0.30 | OC25 | 0.85 |
| 2N3053 | 0.25 | A8Y28 | 0.25 | OC26 | 0.25 |
| 2N3054 | 0.60 | A8Y29 | 0.30 | 0 O 28 | 0.60 |
| 2N3055 | 0.75 | ASY54 | 0.85 | OC29 | 0.60 |
| 2N3133 | 0.80 | A8Z15 | 0.70 | OC30 | 0.76 |
| 2N3134 | 0.80 | A8Z16 | 0.70 | OC35 | 0.60 |
| 2N3391 | 0.20 | ASZ17 | 0.76 | OC36 | 0.60 |
| 2N3392 | 0.15 | A8Z 18 | 0.75 | $0 \mathrm{C42}$ | 0.80 |
| 2N3393 | 0.15 | A8Z20 | 0.25 | $0 \mathrm{C44}$ | 0.80 |
| 2N3394 | 0.15 | A8Z21 | 0.40 | OC45 | 0.16 |
| 2N3395 | 0.20 | BC107 | 0.125 | OC70 | 0.10 |
| 2N3402 | 0.15 | BC108 | 0.126 | $0 \mathrm{C71}$ | 0.18 |
| 2N3403 | 0.15 | BC109 | 0.125 | OC72 | 0.85 |
| 2N3404 | 0.85 | BC113 | 0.85 | $0 \mathrm{OC7}$ | 0.80 |
| 2N3414 | 0.20 | BC118 | 0.80 | $0 \mathrm{C75}$ | 0.80 |
| 2N3415 | 0.15 | BC134 | 0.30 | 0 C 76 | 0.80 |
| 2N3416 | 0.25 | BC147 | 0.175 | 0078 | 0.25 |
| 2N3417 | 0.25 | BC148 | 0.15 | 0C78D | 0.80 |
| 2N3702 | 0.12 | BC149 | 0.16 | $0 \mathrm{C81}$ | 0.25 |
| 2N3703 | 0.12 | BC152 | 0.15 | OC81D | 0.16 |
| 2N3704 | 0.17 | BC158 | 0.15 | $0 \mathrm{C83}$ | 0.80 |
| 2N3707 | 0.15 | BC175 | 0.20 | OC139 | 0.80 |
| 2N3709 | 0.12 | BC186 | 0.25 | OC140 | 0.85 |
| 2N3710 | 0.12 | BCY 30 | 0.85 | OC141 | 0.60 |
| 2N3819 | 0.85 | BGY31 | 0.40 | OC170 | 0.25 |
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| 28702 | 0.50 | BCY34 | 0.80 | OC200 | 0.80 |
| 28746 | 0.26 | BCY 72 | 0.80 | OC201 | 0.60 |
| AC118 | 0.15 | BCZ10 | 0.80 | OC202 | 0.65 |
| AC125 | 0.80 | BCZ11 | 0.40 | 00203 | 0.40 |
| AC126 | 0.20 | BD121 | 0.85 | OC204 | 0.40 |

SYNCHROSCOPE TYPE C1-5


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