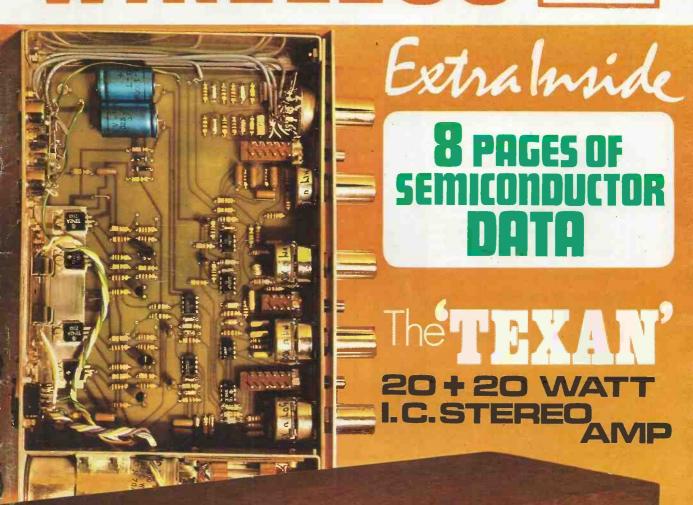
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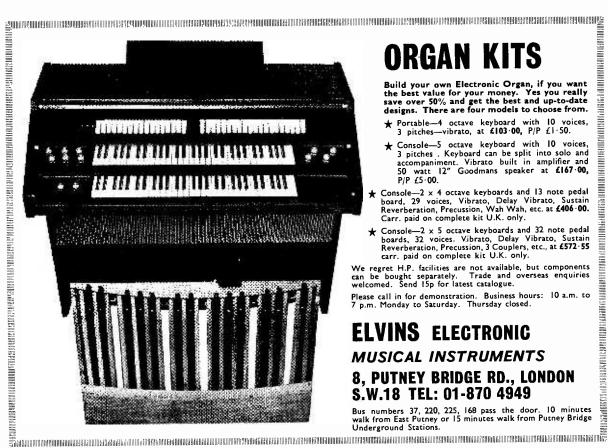
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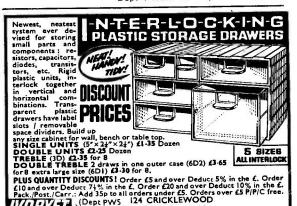
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BC107	0-13	2N I 308-9	0.35
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F.E.T. PRICE

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3823E Field effect Transistors. This is the 2N3823 in Plastic Case, 500 + 13p each: 1,000 + 10p each.

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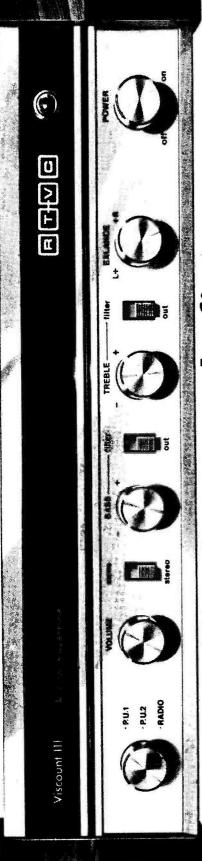
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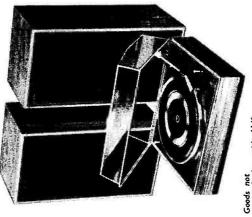
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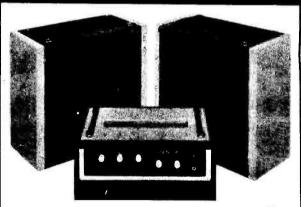
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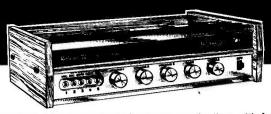
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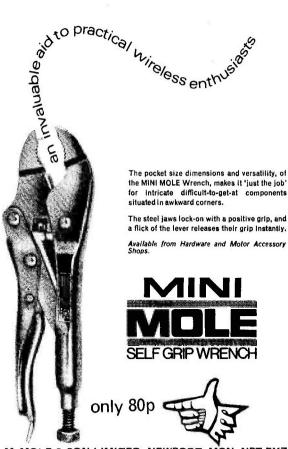
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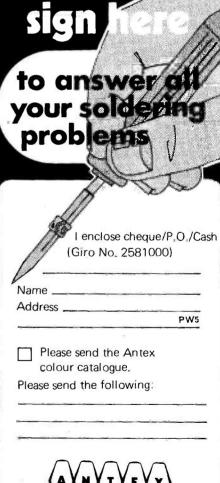
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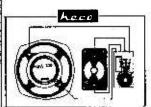
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384	-26	30C18	-61	EB91	-10		-82		-38	UBF80	-84
3₹4	-87	30F5	-64	EBC33	-40		-84	PCL86	-88	UBF89	-82
5U4G	-81	30FL1	-61	EBC41	-54	EY51	-38	PCL88	-65	UCC84	-82
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6AU6	-20	30PL1	-80	ECF80	-81	KT41	•77	PL83	-88	UL41	-57
6BA6	-20	30PL13	-89	ECF82	-26	KT61	-55	PL84	-80	UL84	-80
6BE6	-21	30PL14	-65	ECH35	.55	KT66	-78	PL500	-63	UM84	- 22
6BJ6	-41	35L6GT	-45	ECH42		LN319	-63	PL504	-63	UY41	-39
6BW7	-52	35W4	-25	ECH81	-29	LN329	-72	PM84	-83	UY85	.25
6F14	-40	35Z4GT	-25	ECH83	-40	LN339	-68	PX25	-95	VP4B	•77
6F23	-68	807		ECH84	-86	N78	-87	P¥32	- 55	W77	-48
6F25	-58	AC/VP2	-77	ECL80	-80	PABC80	.84	PY33	-55	277	-22
6J7G	-24	B349		ECL82	-83	PC86	-47	PY81	·25	Transisi	OFF
6K7G	-12	B729		ECL86	-35	PC88	-47		- 25	AC107	-17
6K8G	-17	CCH35		EF39	-38		-42	PY83	·28	AC127	·18
6Q7G	- 85	CY31		EF41	-60	PC97	-89	PY88	-88	AD140	-87
68 N7G		DAF91		EF80	-28			PY800	-84	AF115	-20
6V6G	-28	DAF96	-86	EF85	.28			PY801	-84	AF116	-20
6V6GT 6X4	·28 ·28	DF33 DF91		EF86 EF89	-80		.25	R19	-80	AFI17	-20
6X5G1		DF96		EF91	·26		-40	R20 U25	-56	AF118	-48
	58	DH77		EF92	-80	PCC89 PCC189	·45	U26	-64 -56	AF125 AF127	·17
10P13		DK32	-88	EF98	-65		. 56	U47	-64	OC26	-25
12AT7	-17	DK91	-28	EF183	.28		.28	U47	-56	OC44	-12
12AU7	-20	DK92	-88	EF184	-81	PCF82	-88	U50	.26	OC45	-12
12AX7	-22	DK96	-45	EH90	-85	PCF86	-48	U52	-31	OC71	12
19BG6	G -80	DL35	-40	EL33	-55		-58	U78	-24	OC72	-12
20F2	-67	DL92	-26	EL34	-45		-28	U191	-59	OC75	-12
20P3	-77	DL94	-87	EL41		PCF802	-40	U193	-42	OC81	-12
20P4	-92	DL96		EL84		PCF805	-61	U251	-64	OC81D	-12
25L6G'		DY86		EL90		PCF806	-56	U301 U329	-88	OC82	-12
25U4G		DY87		EL95		PCF808		U801	-66	OC82D OC170	·12 ·28
20 U 4 U	T .01	10101	-23	ET120	-00	LCLSOS	-08	10001	-80	00170	.28

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10 TRANSISTORS. 9 TUNABLE WAVEBANDS, MW1, MW2, LW, SW1, SW2, SW3, TRAWLER BAND. VHF AND LOCAL STATIONS AND AIRCRAFT BAND

Built in Perrite Rod Aerial for MW/LW. Estractable, chrome plated 7 section Telescopic Aerial, can be angled and rotated for peak ahort wave and VHF listening. Push Pull output using 600mw Transistors. Car Aerial and Tape Record Bockets, Switched Earpiece Socket complete with Earpiece. 10 Transistors plus 3 Diodes. 8° × 2° Speaker. Air Spaced ganged Tuning Condenser with VHF section. Volume on/off, Wave Change and Tone Control. Attractive Case in black with silver blocking, Size 9° × 7° × 4°. Easy to follow instructions and diagrams. Parts price list and easy build plans 30p (FREE with parts).

Total building cost

£8·50

P. P. & Ins. 50p (Overseas P. & P. £1)



ROAMER @ @ EIGHT Mk I



7 Tunable Wavebands: MW1, MW2, LW, 8W1, 8W2, 8W3 and Trawler Band. Built in Ferrite Rod Aerial for MW and LW. Retractable chrome plated Telescopic serial for Short Waves. Push pull output using 600mW transistors. Car serial and Tape record sockets. Selectivity switch. Switched earpiece socket complete with earpiece. 8 transistors plus 3 diodes. S' × 2½ Becaker. Air spaced ganged tuning condenser. Volume/on/off, tuning, wave change and tone controls. Attractive case in rich chestnut shade with gold blocking. Size 9 × 7 × 4 in. approx. Easy to follow instructions and diagrams. Parts Price List and Easy Build Plans 23p (FREE with parts).

Total building cost £6-98 P. P. & (Overgeas P. & P. £1)

ROAMER SEVEN MK IV



7 Tunable Wave-bands: MW1, MW2, LW. SW1, SW2, SW2, SW3 and Trawler Band. Extra Medium waveband provides easier tuning of Radio Luxembourg, etc. Built in ferrite rod serial for MW and LW. Retractable 4 section 24in, chrome plated telescopic serial for SW. Booket for Car Aerial. Powerful push-pull output. 7 transistors and 2 diodes, including Micro-Alloy R.F. Transistors 8" × 2; speaker. At spaced ganged tuning condenser. Volume/on/off, tuning and wave change controls. Attractive case with carrying handle. Size 9 × 7 × 4in. Approx. Easy to follow instructions and diagrams. Parts price list and easy build plans 15p (FREE with parts). Earplece with plug and switched socket for private listening, 30p extra.

0

Total building costs £5.98 P. P.&

ROAMER SIX





Sin. Speaker. 8 stages—6 transistors and 2 clodes including Micro-Alloy R.F. Transistors, etc. Attractive black case with red grille, dial and black knobs with pollabed metal inserts. Size 9 \times 5½ \times 2½in. approx. Easy build place and parts price list 15p (FREE with parts). Earpiece with plug and switched socket for private listening 80p extra.

Total building costs £3.98 P. P. &

POCKET FIVE

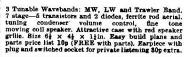
3 Tunable Wavehands: MW, LW, Trawler Band with extended M.W. band for easier tuning of Luxembourg, etc. 7 stages—5 transistors and 2 diodes,

7 stages—5 transistors and 2 diodes, supersensitive ferrite rod aerial, fine tone moving coil speaker. Attractive black and gold case. Size 5∮ × 1∮ × 3∮in. Easy build plans and parts price list 10p (FREE with parts). Earplece with plug and switched socket for private listening 30p extra.

Total building costs £2.23 P.P. & Ins. 21p

TRANSONA FIVE





Total building costs £2.50 P. P. & (Overseas P. & P. 63p)

RADIO

TRANS

8 TRANSISTORS and 3 DIODES

6 Tunable Wave-bands: MW, LW, 8W1, SW2, SW3 and Trawier Band.

and Trawier Band. Sensitive ferrite rod serial for M.W. and L.W. Telescopic aerial for Short Waves. Sin. Speaker. 8 improved type transistors plus 3 diodes. Attractive case in black with red grille, dial and black knobe with polished metal inserts. Size $9 \times 0.5 \times 2$ jin. approx. Push pull output. Battery economiser switch for extended battery life. Ample power to drive a larger speaker. Parisprice list and easy build plans 25p (FREE with parts). Earpiece with plug and switched socket for private listening 30p extra.

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Parts price list and p	olans fo	1¢	and the second
Name			
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	Retail	Comet
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ALBA UA 700	34 50	23 95
ALPHA 212 By Highgate	36 - 75	27 95
AMSTRAD Stereo 8000 Mk 2	27 - 95	16.90
AMSTRAD I.C. 2000	42 95	27 - 95
ARMSTRONG 521	59.00	42.95
DULCI 207	26 . 00	16 - 95
OULCI 207M	32.00	19 - 50
FERROGRAPH F307 Mk. II (cased)	64.00	
FERROGRAPH F307 MI. (Metal case)	60.00	46 95
GOODMANS Maxamp		42 95
GOODMANS Maxamp	54 .00	34 - 95
LEAK Delta 30 (cased)	65.00	48 - 50
LEAK Dolta 70 (cased)	79 - 50	58 - 95
METROSOUND ST20E	39.50	25 95
METROSOUND ST60	70 · 00	49 - 95
PHILIPS RH 591	79 00	56 - 95
PHILIPS RH 590	52 . 00	37-95
PHILIPS RH 580	29.00	19-95
PIONEER SA500A	53-62	38 95
PIONEER SA600	83 75	59-95
PIONEER \$A700	97 - 97	66 - 95
PIONEER SA800	109-21	75.95
PIONEER SA900	133 83	93 95
PIONEER S A1000	137-58	95-95
PIONEER Reverberation 202W	46 35	30 - 95
RANK Rotel 210	34 - 90	24 95
RANK Rotel 310	47:50	31 50
RANK Rotel 610	74 - 50	48 95
ROGERS Ravensbourne	59 - 50	42 50
ROGERS Ravensbourne (cased)	64.00	46 95
	47.50	34 50
ROGERS Ravensbrook Mk, II ROGERS Ravensbrook (cased) Mk.II	52 50	
SINCLAIR 2000		38 50
SINCLAIR Project 60/2 x Z30/PZ5	35·00 23·90	23 - 50
SINCLAIR PROJECT 60/2 x Z50/PZ5	23.90	15-95
DZG/t		
PZ8/trans	34.86	23 25
SINCLAIR PROJECT 605	29.95	21 . 95
SINCLAIR AFU	5 95	4 - 45
SINCLAIR Neoteric	61 . 95	42 . 95
SINCLAIR 3000,	45.00	30 . 95
TELETON SAQ 206	32 - 50	18 · 25
TELETON 307	30.00	19 95
VOLTEX 100w. Stereo Discotheque,		
8 electronically mixed inputs	185 00	139 - 00
WHARFEDALE Linton Amplifler	60 · 00	42 - 50
All take both ceramic and magne	tic car	tridges.
_		

TUNERS

*ARMSTRONG 523 AM/FM	54 . 22	41 - 50
*ARMSTRONG 524 FM	42 - 17	32 - 50
ARMSTRONG M8 Decoder	9 - 50	7 50
*DULCI FMT.7 FM	25.10	17.95
DULCI FMT.7S Stereo	33 - 79	24 - 95
GOODMANS Stereomax	80.07	47 - 95
LEAK Delta FM	75.00	58 50
LEAK Delta AM/FM	89 - 50	68 95
PHILIPS RH 690	45:30	34 - 95
PHILIPS RH 691	85 80	69 95
PIONEER TX500 AM/FM	66.09	53 95
PIONEER TX900 AM/FM	147 15	116 .00
RANK ROTEL 320	54 - 94	39 95
ROGERS Ravensbourne chassis	59 63	45 50
ROGERS Ravensbourne in teak case		
BOCERS Ravellabourne in teak case	64 45	48 95
ROGERS Ravensbrook chassis	43 37	34 95
ROGERS Ravensbrook (cased)	49 - 39	38 95
SINCLAIR 2000	45.00	34 · 95
SINCLAIR 3000	45.00	34 - 95
SINCLAIR Project 60 tuner (stereo)	25.00	19 25
TELETON GT 101	43.93	30.50
All above Tuners are complete	with Mi	×

All above Tuners are complete with MPX
Stereo Decoder except where starred.

TUNER/AMPLIFIERS

AKAI AA 8500	235 · 00	182 95
AKAI 6600	149 - 50	114-95
AKAI 6300	129 50	94 50
AKAI 6200	89 - 50	73.95
ARENA 2600	111 30	69 - 95
ARENA T9000	303 - 45	285 00
ARMSTRONG M8 Decoder	9 · 50	7 50
ARMSTRONG 525	92 · 77	69 95
ARMSTRONG 526	105 - 72	79 95
GOODMANS Module 80, 35w. RMS	91 - 59	68 - 50
GOODMANS Module 80 Compact	165.00	129 - 00
GOODMANS Module 110 FM/MW/		
LW/SW 100W RMS	135.00	108 - 00
LEAK Deita 75	165.00	128 - 95
MIDLAND 19/542	49 58	37 · 50
PHILIPS RH 790	134 - 00	77-95
PIONEER SX770 AM/FM	139 - 23	109 - 95
PIONEER SX990 AM/FM	167-83	129 95
PIONEER 440	104 - 18	82 95
ROGERS Ravensbrook Chassis	99 39	81 - 95
ROGERS Ravensbrook (cased)	108 43	87 95
ROTEL RX150	69.90	53 50
TANDBERG 1171	110 - 00	82 - 50
TANDBERG TR200	99 00	82 95
TELETON F2000	51 - 50	28 - 50
TELETON 10AT1 150w. RMS	160 - 00	94 - 00
TELETON TFS50	79 65	53 95
TELETON TFS50 LA MW/LW/FM	86 - 64	61 95
TELETON CR55	125 - 26	61 95
WHARFEDALE 100-1	139 . 00	92 - 95



All the above Tuners and Tuner/Amplifiers take both ceramic and magnetic cartridges except Teleton F2000 which takes ceramic only. All include MPX Stereo Decoder with the exception of Armstrong where decoder is extra as listed.

Rec. Retail Comet

· ·	Rec. Retail	Comet
	Price	Price
CARTRIDGES		
AUDIO TECHNICA AT66	6.47	4.50
GOLDRING G850	6 27	3 95
GOLDRING G800	12:55	5 95
		10.45
GOLDRING G800E	25.10	
GOLDRING GOOD SUPER E	25.10	14.75
*GOLDRING CS90 Stereo	5.02	4 - 25
*GOLDRING CS91/E	7 53	6 · 35
EMPIRE 1000ZE/X	60 · 34	46 95
EMPIRE 999VE/X	43 01	33 95
EMPIRE 999TE/X	25 69	19 95
EMPIRE 999SE/X	20 31	15.95
EMPIRE 999E/X	15+89	12 30
EMPIRE 999/X	12 - 55	8.85
EMPIRE 909E/X	12 36	9.60
EMPIRE 909/X	9.26	7.25
EMPIRE SOEE/X	9 56	7 . 50
ORDIT Wagnetic NM 22	Special Pri	
ORTOFON SLISE	28.62	22 95
ORTOFON 2 X 15K Transformer		4 95
ORTOFON M15E	29.08	23 50
SHURE M3DM	6-10	4 - 25
SHURE M31E		8 - 69
SHURE M32E		8.00
SHURE M32-3		7 80
SHURE M44-5		5 · 75
SHURE M44-7	7.90	5.50
SHURE M-44C		5 · 50
SHURE M44E	8 60	5 - 95
SHURE M55E	9 . 70	6 - 75
SHURE M75G	14 - 70	9 95
SHURE M75-6		8 45
SHURE M75EJ		10.75
SHURE M75E		12.95
SHURE M75E/95G		14.95
SHURE V15-11		27 - 50
Starred cartridges above are cera		
magnetic.	mic. An ou	iers are
magnetic.		
PICKUP ARMS		
GOLDRING Lenco 75	13.98	9 · 50
GOLDRING Lenco L69		6.50
		24 - 90
SME 3012 with S2 Shell,	. 35 42	26 · 45
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GOLDRING 705/P with G850

GOLDRING GL75 with G800

THORENS 150 AB complete with TX11 cover Shure M55E cartridge

GARRARD	SP25	Mk III	15 85	9 - 95
GARRARD	SL65	В	20 - 48	13 - 40
GARRARD	SL95	В	49 23	32 90
GARRARD	401 .		39 26	27 - 40
GARRARD	SL72	B	31 . 83	23 90
GARRARD	Zero	100	56 - 25	42 - 95
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SP25, SL55, SL65BS	pecial Prin	ce 3:60
GARRARD 40B	. 13 34	10.95
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GOLDRING GL69P Mk. II	. 36 - 21	26 95
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GOLDRING GL75P	. 48.76	35 95
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for G99	. 11.05	9.75
GOLDRING G99	. 26.89	19 90
GOLDRING G101P	. 28 · 68	22 95
GOLDRING GL72 Chassis	. 28 68	22 95
GOLDRING GL72P	. 37 · 29	29 95
GOLDRING GL85	. 62 · 74	47 - 95
GOLDRING GL85P	72 78	55 - 95
GOODMANS 3025	. 37 - 74	19.95
LEAK Delta	. 69.50	59.95
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McDONALD 610	. 19 28	14 50
McDONALD HT 70	. 21.22	15.45
McDONALD HT70 inc. plinth & cor	v. 30 68	22 25
Base and Cover for MP 60 and	d	
610 S	pecial Pri	
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PHILIPS 308 transcription unit com	1-	
plete with base and cover		28 · 95
PIONEER PL12A with base & cover	er 48.34	36 25
THORENS TX25 cover	. 8 52	6 60
THORENS TD125	. 76 03	59 .45
THORENS TD125AB	. 115.57	93.95
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THORENS TD150A Mk. II	. 44.40	34 - 25
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and cover and Shure M44/7 cartr	idge 35.00	27-95

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B. and W DM1 (per pair)	76 94	60 95
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CELESTION Ditton 15	38 47	27 . 25
CELESTION Ditton 25	65 00	44 - 50
CELESTION Ditton 44	54 · 00	39 - 50
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GOODMANS Havant (per pair)	56.00	43.95
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GOODMANS Mezzo 3	34 - 00	22.75
GOODMANS Magnum K2	44.00	29 - 75
GOODMANS Dimension 8	69 · 00	47 - 95
GOODMANS DIN 20NT kit	12 · 54	9 · 95
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tem (pair)	16 - 10	12.75
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KN800 3-speaker System	15.90	10.95
KN1100 4-speaker System	20 · 40	14 95
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KN2100 3-speaker System	30.60	20.95
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LEAK 600	49.50	34 · 50
METROSOUND HFS 103 (pair)	30.05	23 95
METROSOUND 202	21 . 50	15 25
METROSOUND Duplex 15	32 00	23 95
METROSOUND Duplex 25	52.00	36 · 95
PHILIPS RH 411 (pair)	21.20	16 . 95
PHILIPS RH 481 (pair)	21 · 20	16 - 95
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PHILIPS 496	38 - 00	30 · 95
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GOODMANS Axiom 201	17.86	12-25
GOODMANS Axiom 401	9.86	7.25
	5 · 35 5 · 83	3.95
GOODMANS Audiom 10P	5.83	4.50
GOODMANS Audiom 12P	12·37 20 00	9·25 14·75
GOODMANS Audiom 15P	34.00	14:75
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GOODMANS Trebax 100	6.90	4.95
GOODMANS Midax 650	12.95	8 · 95
GOODMANS Attenuator GOODMANS Crossover Networks	3.55	2.50
COODMANS Crossover Networks	9.75	7.15
XO/950/5000	9.75	1.12
XO/950	7.40	4 95
GOODMANS Crossover Networks		
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WHARFEDALE 8in.Bronze/RS/DD	4.82	3·45 6·70
WHARFEDALE Super 8/RS/DD	8 · 19 11 · 20	10.50
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Transformer	0.84	0.70
TO CHESTO CVCTTMC COM		
ALBA UA552 AMSTRAD Stereo 1000 DANSETTE Consort Stereo DECCA Sound 613 DECCA Sound 1204 DECCA SOUND 1204 DECCA A (3) DECCA 4 (3)	45.86	35 - 50
AMSTRAD Stereo 1000	48.00	34 - 95
DANSETTE Consort Stereo	33·74 70·14	29 95
DECCA Sound 613	70·14 66·55	55 · 95 54 · 95
DECCA Sound 614	89.44	79.05
DECCA Compact 3	130 - 00	72 · 95 105 · 00
DECCA Compact o	59 · 50	48 · 95
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FERGUSON 3451 with radio	99·50 44·50	79·95 34·95
DECCA 403 DECCA 403 ELIZABETHAN LZ101 FERGUSON 3450 with radio FERGUSON 3451 with radio FIDELITY UA2 Music Master FIDELITY UA1 Music Master with	44.00	34.93
radio	107-00	81 - 95
GOODMANS Module 80 Compact		
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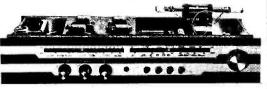


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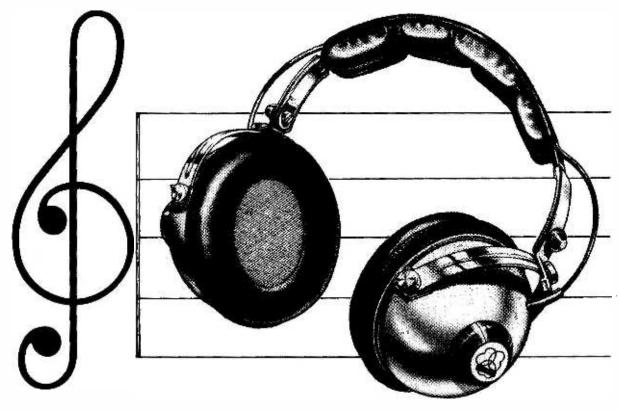
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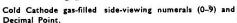


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BP01 = BN7401	0.15 0.14 0.12	BP90 = 8N7490	0.67 0.64 0.58
BP02=8N7402	0.15 0.14 0.12	BP91 = 8N7491AN	0.87 0.84 0.78
BP03 - BN7403	0.15 0.14 0.12	BP92 = BN7492	0.67 0.64 0.58
BP04 = 8N7404	0.15 0.14 0.12	BP93 = 8N7493	0.67 0.64 0.58
BP05 = 8N7405	0.15 0.14 0.12	BP94 = 8N7494	0.77 0.74 0.68
BP07 - BN7407	0.18 0.17 0.16	BP95 = 8N7495	0.77 0.74 0.68
BP08 = 8N7408	0.18 0.17 0.16	BP96 = 8N7496	0.77 0.74 0.68
BP09 = 8N7409	0.18 0.17 0.16	BP100 = 8N74100	1.75 1.65 1.55
BP10 = 8N7410	0.15 0.14 0.18	BP104 = BN74104	0.97 0.94 0.88
BP13-SN7413	0.29 0.26 0.24	BP105 - BN74105	0.97 0.94 0.88
BP16-SN7416	0.48 0.40 0.88	BP107 = SN74107	0.40 0.38 0.36
BP17=8N7417	0.48 0.40 0.88	BP110 - BN74110	0.55 0.58 0.50
BP20 = 8N7420	0.15 0.14 0.12	BP111=8N74111	1.25 1.15 1.00
BP30 = 8N7430	0.15 0.14 0.12	BP119=8N74118	1 00 0 95 0 90
BP40 - SN7440	0-15 0-14 0-19	BP119 - 8N74119	1 35 1 25 1 10
BP41 = 8N7441	0.67 0.64 0.58	BP121=8N74121	0.67 0.64 0.58
BP42 = 8N7442	0-67 0-64 0-58	BP141=8N74141	0.67 0.64 0.58
BP43 = 8N7443	1 95 1 85 1 75	BP145 = 8N74145	1 50 1 40 1 80
BP44 = 8N7444	1 95 1 85 1 75	BP150 = 8N74150	1.80 1.70 1.60
BP45 = 8N7445	1 95 1 85 1 75	BP151 - SN74151	1.00 0.95 0.90
BP46 = SN7446	097 094 088 097 094 088	BP153 = 8N74153	1.20 1.10 0.95
BP47 = 8N7447 BP48 = 8N7448	097 094 088 097 094 088	BP154 = BN74154	1.80 1.70 1.80 1.40 1.80 1.20
BP50=SN7450	0.15 0.14 0.12	BP155 = 8N74155 BP156 = 8N74156	1.40 1.80 1.20
BP51 = 8N7451	0-15 0-14 0-12	BP160 = BN74160	1.80 1.70 1.60
BP53 = 8N7453	0.15 0.14 0.12	BP161=8N74161	1.80 1.70 1.60
BP54 - 8N7454	0.15 0.14 0.12	BP164 = BN74164	2.00 1.90 1.80
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BP60 = 8N7460	0.15 0.14 0.12	BP181 = 8N74181	2 75 2 60 2 40
BP70 = 8N7470	0.29 0.26 0.24	BP182 - 8N74182	0.97 0.94 0.88
BP72 = BN7472	0.29 0.26 0.24	BP190 = 8N74190	8 50 8 25 8 00
BP73 = 8N7473	0.87 0.85 0.82	BP191=8N74191	3-50 3-25 8-00
BP74=8N7474	0.87 0.85 0.82	BP192-8N74192	2.10 1.95 1.75
BP75=8N7475	0 47 0 45 0 42	BP193 = 8N74193	2.10 1.95 1.75
BP76 - SN7476	0.48 0.40 0.88	BP195 = 8N74195	1 10 1 05 0 95
BP80 - SN7480	0.67 0.64 0.58	BP196=8N74196	1.80 1.70 1.60
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BP 702-72702	D.I.L.	14	G.P. OP Amp Wide			
			Band)	53p	45p	40p
BP 70972709	D.I.L.	14	High OP Amp	53p	45p	40p
BP 709P-4A709C	TO-5	8	Righ Galn OP Amp	58p	45p	40p
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BP 741—72741	D.I.L.	14	High Galn OP Amp			
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	50p	UIC48 = 5 × 7448N	50p	$UIC83 = 5 \times 7483N$	60p
	50p	UIC50-12×7450N	50p	$UIC86 = 5 \times 7486N$	50p
$UIC04 = 12 \times 7404N$	50p	$UIC51 - 12 \times 7451N$	50p	$UIC90 = 5 \times 7490N$	60p
	50p	$UIC53 = 12 \times 7453N$	50p	$UIC91 = 5 \times 7491N$	60p
UIC10=12×7410N	50p	$UIC54 = 12 \times 7454N$	50p	$UIC92 - 5 \times 7492N$	60p
UIC13 = 8 x 7413N	50p	$UIC60 = 12 \times 7460N$	80p	$UIC93 = 5 \times 7493N$	50p
$UIC20 = 12 \times 7420N$	50p	$UIC70 = 8 \times 7470N$	50p	$UIC94 \Rightarrow 5 \times 7494N$	50p
UIC40 = 12 x 7440N	50p	$UIC72 = 8 \times 7472N$	50p	$U1C95 = 5 \times 7495N$	50p
UIC41 = 5 × 7441 AN	50p	$UIC73 = 8 \times 7473N$	50p	$UIC96 = 5 \times 7496N$	50p
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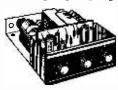


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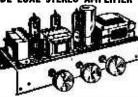


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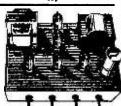
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C	1/8W	5%	4-7-470K	E24	i	0.8	0.7
C	1/4W	10%	4-7-10M	E12	1	0.8	0.7
C	1/2W	5%	4·7-10M	E24	1.2	1	0.9
C	1 W	5%	4-7-10M	E12	2.5	2	1.9
MO	1/2W	2%	10-1M	E24	4	3.5	3
ww	1W	$10\% \pm 1/20 \Omega$	0.22-3.9	E12	7	7	6
ww	3W	5%	12-10K	E12	7	7	6
ww	7W	5%	12-10K	E12	9	9	8

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19p | ACY40

19p | ACY41

26p | ACY44 29p BD132 22p BD135 16p BD136 17p BDY20 17p BF115 1N3754 1N5399 1N5402 17p BF115 31p BF167 19p BF173 23p BF177 20p BF194 21p BF195 21p BF254 63p BF255 1844 18940 2N696 2N697 2 N 706 | 129 | ACY21 | 21p | BF198 | 15p | 16p | ACY22 | 21p | BF296 | 14p | 29p | ACY39 | 21p | BF296 | 14p | 29p | ACY340 | 17p | BFX256 | 15p | 19p | ACY44 | 13p | BFX86 | 22p | 26p | ACY44 | 31p | BFX86 | 23p | 26p | ACY44 | 31p | BFX86 | 23p | 26p | AD142 | 50p | BFX88 | 26p | 33p | AD149 | 50p | BFX88 | 26p | 33p | AD149 | 50p | BFX88 | 26p | 33p | AD150 | 50p | BFY56 | 23p | 23p | AD150 | 50p | BFY56 | 23p | 25p | AF114 | 24p | BFY56 | 23p | 25p | AF114 | 24p | BFY52 | 23p | 25p | AF116 | 22p | BY288 | 18p | 34p | AF117 | 22p | BY288 | 18p | 34p | AF118 | 32p | BY288 | 38p | AF124 | 24p | C407 | 17p | 25p | AF125 | 24p | C762 | 19p | 19p | AF126 | 22p | EA403 | 10p | 35p | AF127 | 22p | EA403 | 10p | 35p | AF127 | 22p | EA403 | 10p | 35p | AF127 | 22p | EA403 | 10p | 35p | AF127 | 22p | EA403 | 10p | 35p | AF127 | 22p | EA403 | 10p | 25p | AF125 | 24p | C762 | 13p | 13p | AF126 | 22p | AF125 | 22p | 2N 706 2N 930 2N 1132 2N 1302 2N 1303 2N1304 2N1304 2N1305 2N1306 2N1307 2N1309 2N1613 2N1711 2N2218A 2N2219 2N2219 2N2270 2N2369A 2N2483 2N2484 2N2646 2N2904 2N2904 2N2905 2N2905A 2N2924 2N2925 2N2926 2N3053 2N3054 2N3702 2N3703 2N3706 2N3706 2N3706 2N3709 2N3711 2N3711 2N3794 2N3819 2N3820 2N3904 2N3906 2N4036 2N4058 2N4060 2N4061 2N4126 2N4284 2N4284 2N4289 2N4291 2N4291 2N4991 2N6457 2N6457 40250 40250 40251 40361 40362 40602 AC107 AC126 AC127 AC128 ACI41H ACI41HK

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cartridge with diamond stylus

PAIR OF STANWAY II Speaker

Units
Special Total Price
Carr. £1·50

Terms: Deposit £13 and 9 monthly
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★ Super 30 Amplifier (15 + 15 watt)

in veneered housing
Goldring GL69 II Transcription
Turntable on Plinth as illustrated
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Matching as recommended for optimum performance.

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★ TA12 AMPLIFIER
6.5 + 6.5 watt in veneered housing
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Individual Ganged Controls: Bass, Treble, Volume and Balance. Printed circuit construction employing 10 Transistors plus Diodes. Output rating I.H.F.M. Frequency range 20-20,000 c.p.s. Bass Control ± 12db. Treble Control ± 12db. Selector switch for P.U. or Tape/Radio. For loudspeaker output impedances of 3 to 15 ohms. For standard 200-250v. A.C. mains operation. Attractive Black and Silver finished metal facia plate and matching

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Consisting of matched 12in. 11,000 line 15 Watt 15 ohm high quality speaker, cross-over unit and tweeter. Smooth response and extended frequency range ensure surprisingly realistic reproduction. OR SENIOR 15 WATT INCLUDING LEON LINE SPEAKER

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Gives pleasing results with £5-35
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HIGH FIDELITY OUTPUT OF 65 WATTS PER CHANKEL
Designed for optimum performance with any crystal
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11 watts R.M.S. (Continuous) into 15 ohms.

12 watts R.M.S. (Continuous) into 15 ohms.

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19 watts R.M.S. (Continuous) into 15 ohms.

19 watts R.M.S. (Continuous) into 15 ohms.

10 watts R.M.S. (Continuous) into 15 ohms.

11 watts R.M.S. (Continuous) into 15 ohms.

12 watts R.M.S. (Continuous) into 15 ohms.

12 watts R.M.S. (Continuous) into 15 ohms.

13 watts R.M.S. (Continuous) into 15 ohms.

14 watts R.M.S. (Continuous) into 15 ohms.

15 watts R.M.S. (Continuous) into 15 ohms.

16 watts R.M.S. (Continuous) into 15 ohms.

17 watts R.M.S. (Continuous) into 15 ohms.

18 watts R.M.S. (Continuous) into 15 ohms.

19 watts R.M.S. (Continuous) into 15 ohms.

19 watts R.M.S. (Continuous) into 15 ohms.

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11 watts R.M.S. (Continuous) into 15 ohms.

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250-0-250v. 100mA, 6:3v. 4a, 0-5-6:3v. 3a, 22-20
300-0-300v. 100mA, 6:3v. 4a, 0-5-6:3v. 3a, 22-20
300-0-300v. 100mA, 6:3v. 4a, 0.5-6:3v. 3a, 22-20
300-0-300v. 100mA, 6:3v. 4a, 0.5-6:3v. 3a, 22-20
300-0-300v. 100mA, 6:3v. 4a, 0.5-6:3v. 3a, 22-25
425-0-425v. 200mA, 6:3v. 4a, 6:3v. 3a, 5v. 3a, 42-54
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A powerful high quality all-purpose unit for lead, rhythm, bass guitar, vocalists, gram, radio, tape, Peak Output rating Londsgreaker unit optional horizontal or vertical mounting.

**Two extra heavy duty 12in Londsgreakers.

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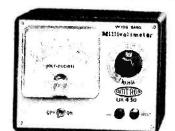


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SOLE AGENT REQUIRED

The Amtron range of products will shortly be available in the U.K. and a sole agent is required to handle distribution. These are just four examples from their extensive range of equipment.

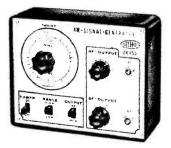


Wide-Band Millivoltmeter

The measurement of small AC voltages has assumed a great importance in the field of electronics. An essential requirement for instruments suitable for making these measurements is that their insertion into the circuit shall not disturb the circuit conditions. The UK430/A can be used for measuring background noise levels, residual ripple in power supplies and the measurement of frequency and gain in amplifiers.

Ratings and Characteristics: Voltage Ranges: 10-30-100 and 300mV, 1-3-10-30-100 and 300V. Decibel Range: -40 to +50dB. Frequency Range: 10Hz to 3MHz. Accuracy: 5%. Input Resistance: 500k ohms 10mV to 3V, 1 Megohm 3V to 30 AF172, Diodes; 4 x AA138.

UK430/A

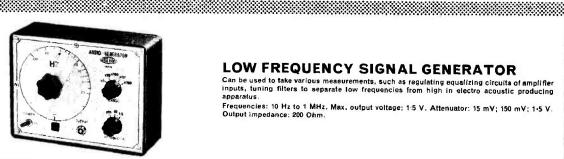


AM SIGNAL GENERATOR

This signal generator is a basic instrument for every technician, useful for AM radio receiver alignment, as well as for a wide range of measurements.

Radio Frequency Output voltage: 100mV. Frequencies: 400 to 900 kHz, and from 950 to 1600 kHz. Radio frequency attenuator: continuous variation, Modulation: internal at i kHz with depth of 30%, possibility of exclusion.

UK455

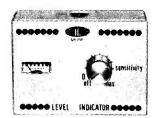


LOW FREQUENCY SIGNAL GENERATOR

Can be used to take various measurements, such as regulating equalizing circuits of amplifler inputs, tuning filters to separate low frequencies from high in electro acoustic producing apparatus.

Frequencies: 10 Hz to 1 MHz, Max, output voltage; 1:5 V. Attenuator: 15 mV; 150 mV; 1-5 V. Output impedance: 200 Ohm.

UK570



Level Meter

This instrument is very useful whenever there is a need to know the level of a signal being fed into equipment. The unit is designed to be used in combination with the UK810 Dynamic Compressor which is provided with an output point for connection to the Level Meter. The instrument embodies a 200µA full-scale-deflection meter, a signal of 5mV amplitude giving full scale deflection.

Ratings and Characteristics: Input voltage: 5mV max. Input Impedance: 47k ohms. Power Source: 8V DC. Transistors; 2 x BC108B, Diodes: OA95

UK 255

Applications are invited from well established companies able to offer distribution facilities throughout the U.K. Please write with full details to:-

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WIRELESS

VOL 48 NO 1

Issue 783

MAY 1972

Up for the Cup!

A T this time of year, all sorts of people are getting worked up on the outcome of that sporting event soon to take place in Wembley Stadium. In between bouts of waving their favours and whirling their rattles, the staff of P.W. have found themselves involved in the fate of another, though less spectacular, silver cup. They have been judging nominations for the P.W. Designer's Trophy for 1971.

In 1970 we launched *Project Autumn*, in which readers were invited to send in original articles on constructional projects within a specified period, the winner being the author of the work considered by the judges to be the best submitted. It was decided, however, to amend the rules slightly for the 1971 Trophy. Entries would be judged not in manuscript form but would be drawn from material actually published between the July 1971 and March 1972 issues.

Nominations would be judged, as before, on originality, ingenuity, construction, presentation, technical accuracy and other associated factors. However, under the revised arrangement, a new factor could be brought into play—call it 'workability', or Factor X. This allows the judges to take into consideration the trouble (or lack of it) in passing the articles through the editorial mill and noting any shortcomings not obvious in manuscript form. This permits a much fairer and realistic assessment to be made. The final results will be published in the next issue.

Although last year's crop of entries brought some new names to the pages of P.W., this year we want to see more and more! It has been suggested that the relative newcomer to technical writing has little chance when in competition with the regular and more experienced authors. There may, of course, be an element of truth in this, because on balance the "published" writer is experienced enough to have learned what is likely to be accepted for publication. On the other hand, P.W. has published many articles by new authors, sometimes at their first attempt.

However, in order to show faith in our stated objective of encouraging new writers, it has been decided that for the 1972 Cup only writers who have never previously had articles published in P.W. will be eligible. Describing the théory, construction and use of a piece of equipment can be a revealing exercise in testing the author's grasp of a subject (and the gaps in his knowledge), with the added incentive of some financial gain and, perhaps, a little glory! Full details next month.

So, keep those articles rolling—you may have just what we have been looking for and (who knows?) you may even win a cup!

W. N. STEVENS-Editor.

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SPECIAL FEATURE

Semiconductor Data Supplement

Part 8 of "Transistor Circuitry for Beginners" is held over.

JUNE ISSUE WILL BE PUBLISHED ON MAY 5th

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NEWS... NEWS... NEWS...

Pacemakers

With surgical care, an electronic module for a long-lived heart pacemaker is assembled at Raytheon Company's Industrial Components Operation at Quincy, Mass. Aerospace reliability techniques are employed in the manufacture of the new devices to pace the beat of a human heart. The new medical aids will be powered by nuclear energy giving them a life expectancy of



10 years instead of two as presently experienced with units powered by mercury batteries. Raytheon is building the modules for ARCO Nuclear of Apollo, Pa., a wholly-owned subsidiary of Atlantic Richfield. The experimental programme is sponsored by the Atomic Energy Commission.

About 50,000 persons now depend upon pacemakers. It is estimated that an additional 5,000 to 10,000 will join their ranks annually as the reliability and longevity of the device is increased.

Mr. Regan, Raytheon's marketing manager, said the electronic circuitry of contemporary pacemakers outlasts the present mercury power sources but with the shift to nuclear energy and its 10-year life, the circuitry had to be upgraded. He said Raytheon had been selected for the task because of the company's experience and expertise in the manufacturing of sophisticated electronic modules. The company's performance with the ultra-reliability of the Apollo programme and compliance with NASA reliability standards were important factors in the selection, he said.

Atlantic Lock Out!!!

Mr. S. Valentine recently wrote to tell us of an amusing incident which occurred when he was doing a bit of S.W. listening one day.

He was tuned in to a QSO between two amateurs—one in the U.S.A. and the other in Scunthorpe. The American asked the "G3" to "let his wife in as it was raining outside." On hearing this, the Englishman was most astonished and enquired what was going on.

It transpired that the English lady of the house had allowed the door to lock behind her, and not being able to make her husband hear her, had contacted one of his pals to get on the air to the American to tell her husband to unlock the door!

1,000W Amp

Crown International of Elkhart Ind. have announced a new high power amplifier capable of providing 1,000 watts r.m.s. into a ohms load. Available from Macinnes Laboratories in the U.K. the M600, is d.c.-coupled throughout although a front panel switch provides for an a.c.-coupled input if required. Power bandwidth at the 1,000 watts level is $\pm 1dB$ from d.c. to 20kHz and THD at this level is below 0.1% throughout the frequency range. At 600 watts THD is below 0.05%. The distortion intermodulation better than 0.1% at all levels from 0.01 watt to 1,000 watts, and the damping factor is greater than 200 from d.c. to 20kHz. Hum and noise are at least 110dB below 600 watts level. The M600 is equipped with its own two speed fan for cooling

"Radio Ear"

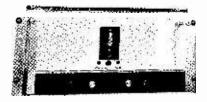
The present generation of mankind may be able to tune in on intelligent civilisations beyond Earth, if Congress appropriates the money to start building the VLA, America's proposed superpowerful radio observatory.

The National Science Foundation, under whose aegis the radiotelescope would be designed and constructed, said that an array of aerials resembling the giant "dishes" used to track spacecraft, would be set up in geometric patterns to pick up and focus signals emanating from the edge of the universe, 13,000 to 16,000 million light years away. The pattern would be an extension of the interferometer technique (placing aerials in a row or at right angles to one another) already widely used in radioastronomy to enhance the strength of incoming radio signals.

A specific location has not yet been chosen for the new telescope, but it would be built in a pattern some 24 miles long at a high-altitude site in the southwestern part of the United States, well away from earth-generated background noise.

It is likely that any new very large aerial would be used, at least part time, in "passive listening" for radio signals from other civilisations which many scientists believe must inhabit planetary systems around distant stars.

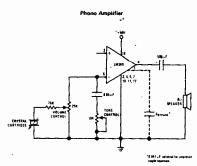
and on the rear panel is a switch offering six positions of operating modes. Weight is 80lb., and price inclusive of import duty, is £730. Macinnes Laboratories Ltd., 71 Oakley Road, Chinnor, Oxon. Tel. Kingston Blount 52061.



The M600 amplifier

NEWS... NEWS... NEWS...

LM380 I.C.



D.T.V. Group Limited announce the LM380 audio power amplifier. The output has both short circuit current limiting and thermal overload current limiting to insure safe operation. Supply Voltage is 22V. Peak Current is 1.3A. Power output is 2W and input voltage is ±0.5V. Storage Temperature is -65°C to 150°C and Operating Temperature 0°C to 70°C. Lead Temperature (Soldering, 10 sec) 300°C. Price is £1.3712 +10p postage and packing. D.T.V. Group Limited, 126 Hamilton Road, West Norwood, London, S.E.27.

Gas Radio

Reader R. Jenkins has told us of a Mr. Jack Hickson who has a Gas-Powered radio.

The radio at Mr. Hickson's home in Hadnock Road, Monmouth, takes its power from a thermo-electric generator which has gas burners. Electricity is generated by heating of copper plates. The generator charges a battery off which the 33-year-old radio is run.

The generator supplies enough electricity to charge accumulator batteries or even a car battery.

The generator was made in 1938 by a Yorkshire firm specially for providing power for battery radios.

But for gas board engineers carrying out North Sea conversion it was something they had never encountered before. They had to take the generator to Cardiff and it took them about a week to change the burners and adapt it for North Sea gas.

Audio Fair

Exhibitors at last year's International Audio Festival & Fair asked for more space in the 1972 show. In order to accommodate them, the organizers have arranged to move the Festival from the Empire Hall to the Grand Hall, Olympia, which will provide an additional 74,000 sq. ft. of space.

The 1972 Audio Fair, sponsored by the Sunday Mirror, will be held from October 23 to 28. It will allow the public to listen to the products of the world's leading manufacturers of sound reproduction equipment, specialist Hi-Fi, and a comprehensive display of accessories.

There will be many special attractions, including more than 75 specially constructed Audio Studios, and a Hi-Fi Theatre offering a daily programme of lectures and concerts.

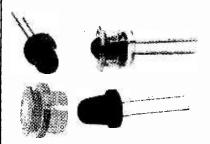
GaAs Light



GaAs Light indicates the state of p.n.p. or n.p.n. logic systems capable of driving at say 15mA at 1.6V or over. For the brightest display 40mA is recommended. The intensity is acceptable at only 10mA. The advantage of these lamps is the low cost (95p) and higher reliability.

Due to the low current and voltage requirement of small battery operated equipment it is now possible to run these GaAs lamps in series with the electronics, so reducing the indicating power to only 16mW. This power is therefore less than the power wasted by even a very small equipment being inadvertently left on. West Hyde Developments Limited, Ryefield Crescent, Northwood Hills, Northwood, Middx., HA6 INN.

Led Lamps



Guest International announce the availability of a new range of solid state lamps using Gallium Arsenide Phosphide Diodes. The lamps are encapsulated in two main forms of package-one with a glass to metal seal and the other a moulded plastic case. Both employ a lens in front of the actual diode to alter the characteristics of the light emission, one type diffuses the light, whilst the other produces a point source each available in different colourations. The lamps combine extremely long life with very efficient operation making them ideal for use as panel lights or status indicators in modern equipment. There are currently 5 versions in the range given the generic title RED LIT series and luminosities as high as 2000FtL at 50mA are offered. Guest International Limited, Nicholas House, Brigstock Road, Thornton Heath, Surrey, CR4 7JA.

Philips Booklet

Philips have published a colour booklet entitled "All about Tape and Cassette recording". The booklet, which sets out the origins of tape recording and describes how magnetic tape works, is available to schools and students on request.

It has been published primarily as further study material for project work.

25,000 copies of the booklet have been printed and copies can be obtained from the Publicity Department, Philips Electrical Ltd., ELA Division, Century House, Shaftesbury Avenue, WC2H 8AS.



▼INCE the orginal Treasure Tracer was built and described in this magazine, development work has continued. The author's own success with the prototype and the enthusiasm of readers has resulted in a considerable amount of experimenting and this article describes the improvements that have been made.

It has not been assumed that all readers will have seen the original article; for this reason much of the information given last time is repeated. This article should enable the new reader to build the project as described as well as enabling the reader who built up the original circuit to up-date his Treasure Tracer. To make matters simpler for these people, the original case has been retained with relatively few changes.

The cost of the original was not high but in fact the component cost for the Mark II will probably be marginally less, though the mechanical and construction side may cost more. The total cost for the Treasure Tracer Mk II should be something under £5 though this will vary depending on source of

supply.

For those who did not see the original article, the principle of operation will be briefly described.

The circuit works in exactly the same way as the original. Two low frequency r.f. oscillators are arranged to beat together. One of these oscillators makes use of a tuned circuit working inside the case; we shall call this the reference oscillator. In operation this remains at one frequency though this is adjustable for "tuning up".

The second oscillator, working at essentially the same frequency, also makes use of a tuned circuit but in this case the inductor is external and becomes the "business end" of the circuit. We shall call this the

search coil and the search oscillator.

If a small part of each of these oscillators is mixed together they will beat with each other. If the difference in frequency is small, this beat note will fall in the audio range. If the reference oscillator is working at say 130kHz and the search oscillator is at 131kHz, a beat note equivalent to that difference will be produced, in this case 1kHz; this audio signal can be amplified and fed to the speaker.

Now the search oscillator makes use of an external coil and if certain materials, especially metals, are brought into the vicinity of this coil, these will alter

THE



The original Treasure Tracer, described in P.W., August 1971, was a very popular feature, judging by your letters and reports from components stockists. Few readers had difficulty in either building or using the metal locator. This article describes an improved version of this project.

HALVOR MOORSHEAD

the value of the coil inductance, which in turn will alter the frequency of oscillation.

A better known example of this change of inductance is the ferrite rod aerial. By moving the coil along the rod the inductance is changed; this is of course done every time a superhet is aligned. A ferrite rod will roughly double the inductance of a coil. Far smaller changes in inductance take place in a metal locator—usually fractions of one percent—but because of the frequency used and the fact that it is compared to a standard frequency, the changes in the audio note will be considerable. Taking the frequency actually used in this circuit, a 0·1 per cent increase (or decrease) in the inductance will result in a 130Hz change in the audio note.

It would be possible to have both frequencies fixed at say a difference of 100Hz but there are so many factors that effect the frequency that it is easier to arrange for one of the oscillators to be adjustable in order that they can be brought close together.

In this circuit we are changing the frequency of the reference oscillator though it could just as well have been the search oscillator.



MODIFICATIONS

In the light of experience and bearing in mind the comments of readers, the following modifications have been made to the prototype:

- 1. The original Treasure Tracer was not mechanically balanced. It was held at the end of the handle and while this worked quite well, it became very tiring to hold during long searches. The Mark II is balanced as can be seen from the photograph; it is easier and more comfortable to hold and there is no hand strain even after several hours.
- 2. The output volume was felt by some readers to be too low. In many cases this could have been due

to inefficient operation of the particular transistors used as the output of the prototype was found to be quite adequate. Even so, the sound level available in the Mark II has been increased and since this could be annoying in certain locations at maximum volume, a volume control has been included. To save extra drilling etc., the on-off switch (a miniature slide type in the prototype) is incorporated with the volume control.

3. The stability of the prototype left a little to be desired. Touching the metal case or the wires running from the search coil to the case caused a change in frequency. This was not too serious in practice once one got used to it but several changes have been made to the construction of the Mark II to improve the stability.

4. The construction of the search coil has been changed to make the windings much more rigid; this has proved to be much better than the principle originally employed. In addition the weight of the "head" has been reduced and this again helps with the balance.

5. Overseas readers (who account for a proportion of the magazine's readership) sometimes had difficulty in obtaining certain parts. This has been borne in mind and the parts should be available world-wide except for one component and the construction of this is described later.

6. Improvements have been made to the oscillator circuits. In the Mark I a Hartley oscillator and a Blocking oscillator were used for the search and reference oscillators respectively. Both of these have been changed to the more stable (and slightly simpler) Colpitt's type. In addition to these advantages, the search coil is now only a single winding (it was tapped in the Mark I) as is the coil used for the reference oscillator. This simplifies construction and reduces the cost slightly.

In addition to the above points a considerable number of other experiments have been tried but have not been incorporated for various reasons. For those who wish to experiment further, these are mentioned; they may give certain leads and possibly prevent the reader following fruitless paths.

The best frequency for operation was carefully investigated. Although the regulations limit operation below 150kHz, experiments were carried out at six widely differing frequencies and the results compared. Since for a given percentage change in inductance there will be a much larger change in the audio frequency using higher r.f. frequencies, it would appear that operating at say 2MHz would be better. In practice other factors start to come into it and ground capacity effects become serious. Bringing the coil towards plain earth causes a considerable change in frequency and even fitting a Faraday Shield only improved this slightly. About six different frequencies were tried but the only result was a confirmation that the original 130kHz or thereabouts is the best. Ground capacity effects are negligible and, although a Faraday shield was tried, this made no improvement and so has not been incorporated.

A meter was tried as the sensing indicator. Although a workable circuit was developed, it was found to be far less sensitive than the ears and very difficult to use. There was also considerable difficulty influency counting circuit. For this reason details are not included.

Several experiments were made with different diameters of search coil. A clear rule seems to apply here. The larger the diameter of the coil, the greater the depth of penetration for large objects but at the same time the larger coils were less sensitive to

small objects close by.

There is certainly room for experiment here. As long as the inductance of the coil remains the same, it doesn't matter at all what size it is, it could be lin. or 3ft. in diameter. The 6in. used here is the best compromise for general purposes; it is good for coins and not too bad for large objects reasonably close by. For those who are prepared to take the trouble, it should be possible to arrange for the coils to be changed depending on the nature of the search.

As a general guide the same length of wire should be used whatever the diameter of the coil, that is about 92ft. of wire. This is only a guide and turns may have to be added or taken off, to allow the frequency to be brought into the right range.

Encapsulating the coil in epoxy resin was tried and although this was an improvement over the Mark I as it prevented movement and consequent frequency changes, it did not help with the Mark II and only added to the weight of the search coil head.

THE CIRCUIT

The complete circuit is shown in Fig. 1. Trl and the components surrounding it form the search coil oscillator and, as can be seen, this connects to the search coil via a coaxial cable. Tr2 and the associated components form the reference oscillator, L2 being a long-wave winding (an inductance of just over 2mH) but fitted with a ferrite slug (not normally supplied). The one used in the prototype is primarily intended as a crystal set coil; a medium wave winding is also fitted but this is ignored.

This oscillator is adjustable by varying VC1 which is a 750pF compression trimmer and the combina-

* components list

Resistors 120kΩ 470kΩ R5 See text 10kΩ R6 470kΩ R7 270kΩ **R8** 39Ω 10kΩ 84 All resistors ¿W, 5% types. VR1 10kΩ log pot with switch

Capacitors

C1 2,000pF ceramic or polystyrene

2,000 F ceremic or polystyrene 0.05 F ceramic or mylar

3,300pF caramic or polystyrene

0 05µF ceramic or mylar C5

C6 27pf polystyrene or mica 27pf polystyrene or mica C7

3,300pF ceramic or polystyrene C8

0-1 F ceramic or mylar C9 F 10V electrolytic C10

50uF 10V electrolytic C11

Semiconductors

Tr4 BC109 or 2N2926 Tr1 28/2926 D1 1N914

Tr2 Tr3 2142326

(Note that any colour of the 2N2926 can be used but the "G" or green is best for Tr3 and Tr4)

Miscellaneous

See text and Fig. 2

Repanco Crystal Set Coil Type DRX1. This must be fitted with a ferrox tuning slug which s not provided.

Transistor output transformer, Type LT700 or

similar—see text.

LS. Miniature 8Ω loudspeaker—see text Earphone jack, 3.5mm type with break switch; Battery, PP3, 9V; Case, 61 × 21 × 11 in. in aluminium favailable from H. L. Smith Ltd., 287/9 Edgware Road, London W.2, price 60p including postage); Coax cable; Connecting block; Aluminium (or copper) tube, 5ft; Plywood for former.

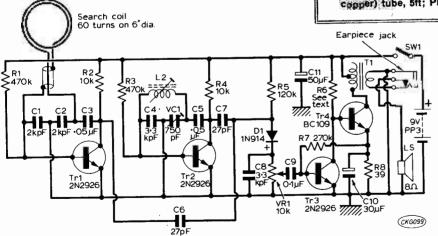


Fig. 1: The complete circuit of the Mark II. Although very similar to the original as far as operation is concerned, this is more stable and has a greater output.

tion of the variable core in the coil and VC1 enables this tuned circuit to be varied over a considerable range. Even if the search coil oscillator is way off the intended frequency it will be possible to tune the reference oscillator to it.

From the collectors of both oscillator transistors low value capacitors C6 and C7 feed to the detector diode which, being a silicon type, requires a standing d.c. bais provided by R5. The output from this is fed to the volume control VR1 and is smoothed by C8 which removes nearly all the r.f. component. The output of the volume control feeds to the simple amplifier which uses Tr3 and Tr4. R6 has not been given a value as it should be selected individually; its value will depend on the actual transistors used. For initial setting up a 22kΩ resistor may be used, this will certainly work. Once operation of the main stages has been established, the correct value should be found by experiment. It should be chosen for maximum output compatible with reasonable current consumption (under 20mA).

Some readers had difficulty in obtaining the high impedance speakers used in the Mark I and for this reason the Mark II uses a standard 8Ω miniature loudspeaker. This necessitates using a transistor output transformer. Normally only push-pull types are available, that means that the primary is centre tapped, in this case only one half of the primary winding is used. The use of earphones is also easier as the impedance of 8Ω is far more common than the 800 used in the Mark I. For those who have high impedance speakers (anything from 35Ω to 80Ω) these can be used, in which case they go directly between the collector of Tr4 and the positive line. The fitting of an earphone socket for this will have to be worked out if required. R6 will be a different value if a high impedance speaker is used.

The whole circuit is operated by a PP3, 9V battery which will last a considerable while. The current consumption will depend on the value of R6 but will probably lie in the range 10-20mA.

For those who have the original Treasure Tracer article, it will be seen that although very similar in many aspects, this circuit is rather simpler as well as being a bit better. The choice of a BC109 transistor for Tr4 rather than the 2N2926 used previously is not important and a 2N2926 can be used here if required.

CONSTRUCTION

The construction of the Treasure Tracer Mark II can be regarded in two distinct sections; the electronics and the mechanics. It does not matter which is tackled first.

The search coil is wound on a wooden former cut from a piece of ¹4in. (6mm) plywood, see Fig. 2. This is circular in shape with a cut-out to reduce the weight although this is optional; it is a simple matter with a fret-saw or jig-saw but difficult otherwise. A groove should be cut into the edge of this circle to hold the wires. Many methods can be used to do this, the simplest way is to spin it in an electric drill and use a rasp to cut the groove. The search coil need not even be round, it could be square or triangular and many other methods of construction can be employed but the described shape has proved very successful. When this groove has been cut, two very small holes should be drilled diagonally from the groove to the top of the former as shown. Where these emerge, a connecting block can be screwed on to act as the winding anchoring point. Sixty turns in all are required; these should be neat and taut and once finished a band of tape can be wound around as shown in the photograph. An alternative would be to saturate the windings with a glue to hold them firmly in place. By sinking the windings in a groove they are protected to a considerable extent.

The handle of the device is made from aluminium tubing which is available from a number of sources. If difficulty is experienced in obtaining this it can be made from copper tubing with the same diameter, a plumber should be able to supply a suitable length. Five feet was found to be about the right length but this will depend on the height of the user.

The handle is fitted to the search coil former with a simple bracket, bolt and wing-nut as shown in Fig. 2. The connecting wire from the search coil to

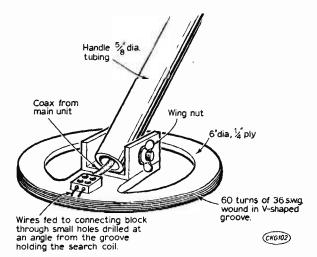
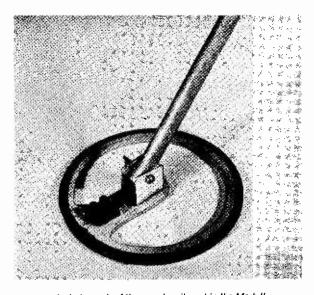


Fig. 2: The construction of the search coil. This design has been found to be excellent but as long as the inductance of the coil remains the . same, any arrangement may be used.



A photograph of the search coil used in the Mark II.

the main circuit is a coax cable running up the tube. At the lower end this is fitted to the connecting block. The use of coax ensures that there will not be loose wires flopping about inside and that the capacity will remain constant. Coax is more rigid than twin wires and is also better for that reason.

The electronics is built into a aluminium case, available from the suppliers mentioned in the components list. This will have to be drilled to take the speaker and various other fittings. A drilling diagram is not shown as the siting will depend on the components used and in any case the siting is fairly obvious. A series of small holes can be drilled for the speaker or a large hole (1¹2in. diameter at least) which can be fitted with some form of grille.

The majority of the components are mounted on a piece of Veroboard and the layout of this is shown in Fig. 3. Veropins are used for all the take-off points as this leads to a nice, neat appearance. The coil L2 is mounted directly onto the board as shown. Four corner mounting holes should be drilled for later fitting.

A number of breaks are needed in the copper strip of the Veroboard and these are indicated in Fig. 3. There is nothing complex about the wiring. R6, which will have to be found by experiment, has plenty of room around it. The easiest way of experimenting with this is to initially wire in the recommended $22k\Omega$ resistor with fairly long leads. Once the general operation has been established this can be cut out leaving wires sticking out. These can form the anchoring point for the replacement.

Figure 4 shows the remainder of the wiring and also the anchoring of the coax cable from the search coil. All connecting points are coded with letters and no difficulty should be experienced in this part of the construction.

Fig. 3: The components layout on Veroboard. Veropins should be inserted at the points marked A to H.

TESTING

Once completed the unit can be tested. If all is well there will be a slight noise from the speaker which will be increased greatly by dabbing a finger onto the VR1 contact which runs to point A on the board. Set VC1 to roughly half compression and adjust the core of L2. A strong whistle should be heard at one point and this will be found to be tuneable by VC1 as well. Other lower level whistles may

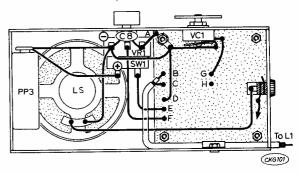
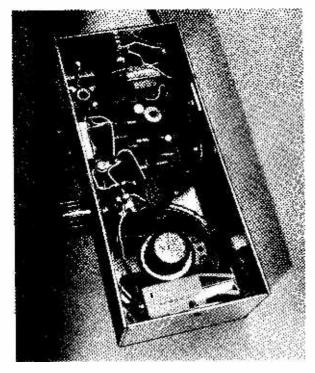


Fig. 4: The wiring inside the case. VC1 must be sited so that it does not foul the coil L2.

be heard, these are due to the oscillators beating with radio stations. In some cases it is possible to make the unit operate using only these but it will not be nearly as loud or as good. The level of the main beat note will be much louder and cannot be missed.

Once the beat note has been established, the Treasure Tracer is ready for use. We have deliberately avoided making any claims about the depth to which it will find objects. Commercial models frequently make wild claims which cannot be substantiated and we do not want to follow this example.

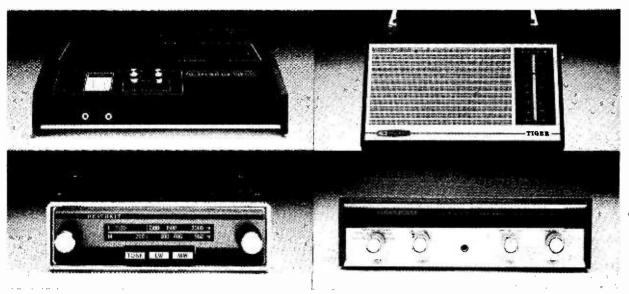


An internal view of the Treasure Tracer Mark II.

To conform with the Wireless Telegraphy Act (1949) a licence is required to use the Treasure Tracer described here.

Under Section 1(1) (Pipe Finder Licence) the band 16 to 150kHz can be used for equipment of this type.

A licence for five years costs 75p and can be applied for on a form obtainable from the Ministry of Posts and Telecommunications, Waterloo Bridge House, Waterloo Road, London S.E.1.



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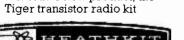
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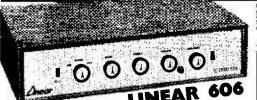
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between 3 and 15 ohms. Stereo/Mono Switch. ★ Input Selector Switch. Solid State Circuitry.



0-200-230-250v. 50 Hz A.C. mains operation. Inputs for magnetic or Ceramic Pickup, Tape or Radio Tuner.

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TECHNICAL DETAILS

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Sensitivities Mag. P.U. 3-5 m.v. into 47K ohm R.I.A.A. Ceramic P.U. 35 m.v. into 100K ohm. Tape Amp. 100 m.v. into 100K. Radio Tuher 400 m.v. into 400K ohm, Crosstalk 53 dB.

Hum and Noise-75 dB min. vol. -65 dB max. vol.

Total Harmonic Distortion 0.1% at 1watt into 15 ohms.

Output (per channel) 6.5 watts I.H.F.M.

TECHNICAL DETAILS

Frequency Range 20 Hz to

Output (per channel) 5 watts

Bass Control ± 12 dB at 60 Hz.

Treble Control ± 14 dB. at

1.H.F.M.

14 KHz.

0-200-250v. 50 Hz A.C. mains operation

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DEPTH OF OPERATION

In the original article we showed exactly what we found and at what depths. Generally speaking coins can be detected at ranges of up to 4 or 5in. but this also depends on the size. Large objects can be detected at distances up to about a foot but the sensitivity falls off considerably over that distance.

Do no expect to become an expert as soon as you start to use the Treasure Tracer. The Mark II is considerably more stable than the Mark I and there will be fewer false signals; even so it takes an appreciable amount of practice before the small changes in the audio frequency are noticed. This point cannot be over-emphasised.

Using the Mark II one person was able to locate a 12p coin at 4in. every time, while another, who had never tried it before, had difficulty in finding a 50p coin at 2in. This illustrates the importance of practice.

The beat note is adjustable for either a rise or fall in frequency when a metal object is approached. The author prefers a falling note but this is only personal preference. It is usually better to work with a low frequency audio note-say 100Hz-as a 50Hz change at this frequency is very noticeable; at 1kHz the same change would not be so obvious.

FREQUENCY LOCKING

If the two notes are very close together, the two oscillators will lock together. The Mark II is slightly better than the Mark I in this respect and reducing the value of C6 and C7 will lower this locking frequency even further but no advantage was found in bringing it much lower. The prototype Mark II locked at about 20Hz.

VC1 has considerable range and it has been found better to arrange for it to be used at near maximum compression, compensating for this by readjusting the core of L2. This prevents a change in frequency due to the leaves of VC1 moving which they tend to do when there is little compression. A possible improvement would be to replace VC1 with a much lower value-say around 100pF (compression trimmers in this range are available) and to wire fixed capacitors in parallel with it to obtain the correct value. This will mean that VC1 will have a narrower range, but still around the correct frequency, at the same time as reducing the likelihood of instability.

L2, as we have said, is a long-wave winding of a crystal set coil, the only modification being to fit a ferrite core (the former is already tapped to take this). It is possible, if tedious, to wind one's own coil on a ¹4in, outside diameter former; 350 turns of fine wire (39s.w.g. is used by the makers) pile wound will give the correct inductance.

One point must be repeated from the previous article. Users of metal locators are not very popular with many archaeologists because of the highly irresponsible way in which some people use theirs. A local archaeological society may be happy to use your services under their supervision but others may not welcome you. Note that there are heavy penalties for using such a device in areas scheduled as being of historical interest without permission.

We regret that issues of Practical Wireless, August 1971, which described the Mark I are not available.



ANY people have expressed a mixture of interest and scepticism about the 50F (fifty IVI Farad) "capacitors" which are contained within ¹₃rd of a cubic inch. These are manufactured by Gould Ionics in the U.S. and are marketed in the U.K. by Lyons Instruments at Hoddesdon. Note: the price of these units for one-off is around £25 each and the company may well only supply the professional user.

How does your radio shop test those transistor radios and televisions? Bet they have an engineer prodding about with a multimeter. Of course, knowing exactly where to prod and why is the most important part.

In industry, it would take far too long to do this on the end of a fast-moving production line so other techniques have had to be employed. Automated test gear is a highly sophisticated subject, but it is becoming more commonly employed.

One piece of gear which does this has been introduced by Honeywell Limited. Designated the "Swift" digital tester, it is a portable equipment and is so simple to operate that even an unskilled person can use it successfully. It can be used for testing digital printed circuit boards when they are fully assembled. Amazing thing is the speed at which the tests can be carried out. The equipment can test at the rate of a million tests every second which is just a trifle faster than the average service engineer. A p.c.b. with some 64 connectors may be plugged in and tested and should a fault arise, this is indicated by a light which shows at which end connector the fault is.

Cable-Scan is another useful little test gadget marketed by Thomas & Betts International Inc. located at Cowcross Street, London. Imagine you had a bundle of wires to sort out. Which are the signal wires, which are earth etc? Well, Cable-Scan will tell you.

A clever idea is that the operator wears a small wrist strap which provides a signal. This is then cunningly transmitted via the operator's fingers so that all he has to do is to touch the wires. A digital readout is provided so there is no chance of an operator making an error through having to write data down-he can see at a glance. Up to 200 signal input circuits are catered for which means that the operator can tackle a mighty big bunch of wires and sort them out without any trouble.

Before you think of rushing out and buying an automated tester and setting up in business as a lightning radio and t.v. repair enterprise, the cost of the Honeywell Swift equipment is around £900. But it is still cheap for large production runs.

The "colonials"

I am writing in hope that you may like to enlighten "us Colonials" about some of your circuits that you print in that great little magazine P.W.

We in Australia cannot obtain quite a few components that you use, mainly coils and substitute transistors. Wouldn't it be possible to give information on coils, such as diameter, turns, gauge of wire, etc.

There is another favour you can do for your readers, don't change P.W., it's O.K. as it is .--R. C. Shambler, (Australia). Editor's Note:

It would not really be practical to give all the information on the coils specified in P.W. However the majority of the larger advertisers in this magazine will supply components to overseas readers and even air mail postage on small items such as coils and transistors is surprisingly inexpensive. (To Australia the air mail postage is often only twice that of internal British mail).

Cassette deck

I agree with V. C. Watts (Bath) "Cassette deck" March '72. I have tried many companies in search of a deck for a home-made system. The only firm I have discovered who at one time did sell decks (made by Philips) has now stopped and according to their letter did not think they would be able to sell them again. Perhaps these and other letters may stir the component firms into action.-A. R. Knight, (Oxford).

Ebuc oidar

Leafing through my March copy of Practical Wireless I chanced upon the article on the Cube Radio. On closer inspection. I found that the photograph on page 998, a general view of the completed receiver, was shown as a mirror image. The lettering on the battery and the position of VR1 show this. Is this because we have to use a battery providmirroring positrons—the images of electrons?-G. L. Manning, (Middlesex).

Alice is not a member of our Staff—it's all done by mirrors— (Editor).

Tools for the job

The purpose of this letter is comment on "Who makes What?" In Feb. issues's "Letters", as it has some bearing on my own experience, although with ancillary tools rather than radio parts.

Last May I got a kit for the "PW-12-12 Stereo", which is now, after some odd troubles working fantastically well. I was determined to get the tools for the which included special soldering irons and bits, the standard 'nipper' type of wire cutting pliers, and long nosed ordinary pliers, I was also determined to get British tools which I did; or thought I did. My comments hereunder refer to the pliers and wire nippers.

They were both on the modern type of display card, the long nosed pliers, with nicely red painted handles I saw were made in England. Quite a number of filing marks here and there on them, and a scratchless finish which looked like chrome plating. These nippers had the name in large headlines of the English company which 'made' them as I thought A London firm. Surdespite the perfect prisingly finish and insulated grip of the 'nippers' there was only 10% difference in the prices, and I thought that British tool makers at last were 'With it!'

When I got home I found that these perfect nippers were made in Japan, as quite properly announced in tiny print, for the English firm, and I felt aggrieved.

So I went to work on my Double-12 amplifier, and exactly a month later I found that the English made pliers had $^{1}_{16}$ in lateral play at the plier points, showing either bad workmanship or bad material. After nine months the Japanese nippers are as when bought and when I'm sewing on a trouser button, I use them in preference to scissors to cut the cotton! Clean as a whistle.

I know this does not refer to radio parts, but such an experience reflects on all British products, and I wonder if this is the reason why it is now apparently expedient not to put the country of origin on the product. -N. H. Hodgson, (Middlesex).

CQ thanks

I should like to thank you for printing my problem in your C.O. column. I have had several replies recently, including one offer from a Mr. Oldknow in Horsley, Woodhouse who is willing to turn a new pulley wheel from scratch on his lathe!-W. R. Kennon, (Derbyshire).

Dreams

I often dream of being a ship's radio operator between the sparkgap and the valve era, with big brass morse keys, large accumulator terminals, 4 pin valves and coils on 2in formers and not 14in. The transistor should never have been invented. Old and sentimental at 26 years of age!

Just in case I start a war as an inexperienced nut perhaps the following details provide my defence. My construction knowledge covers from a crystal receiver to a 9 valve communications receiver.

Operating experience - 42,000 miles of travel.

Operating a.m., s.s.b., c.w. and RTTY links plus Australia with the armed forces, presently employed as GPO Telegraphist.

Transistors are too small and don't leave enough room for any serious experiments, I'll still keep buying old battery valve sets for 50p and £1 and re-making them until they run out of supply and then I'll take up fishing.-W. D. Logan, (Hyde, Cheshire).

Nuvistors

In "Letters" of March, Mr. K. Freeby remarks on the absence of information on nuvistors. I have a copy of Practical Television dated December 1962 in which there is an illustrated article by K. Royal entitled "Practical Nuvistor Circuits". Mr. Freeby may have this magazine if he requires it .- N. A. Hodgson, (84 Lansbury Drive, Hayes, Mid-

We would like to thank Mr. Hodgson for his kind offer and also to thank Mr. Laurence Mason who has kindly sent us some gen on nuvistor valves to pass on to Mr. Freeby-Editor.

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Six speeds are available 500, 850 and 1,100 r.p.m. and 8,000; 12,000 & 15,500 r.p.m. shaft is 1 in. diameter 230/240v Its speed may be further controlled with the use further controlled with the use of our Thyrister controller. Very powerful and useful motor size approx. 2 in. dia. × 5 in. long, Price 88p plus 23p postage and





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How long does it take you to renew a fuse? Time yourself when next one blows. Then reckoning your time at 21 per hour see how quickly our resettable fuse (auto circuit breaker) will pay for itself. Price only \$1 each or

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for portable, car radio or transmitter. Chrome plated—six sections, extends from to 47in. Hole in bottom for fBA screw, 88p. KNUCKLED MODEL FOR F.M. 50p.

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This Tuner is a precision instrument made by the tamour 'Cyldon'
Company for the equally famous
Radiomobile Car Radio. It is a medium wave tuner (but set of longwave coils available, 259 with a frequency coverage 1620 Kc/s
526 Kc/s and intended to operate with an I.F. value of 470 Kc/s.
Extremely compact (size only 21+
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Standard size 14 wafer-silver-plated 5-amp contact standard I" spindle 2" long-with locking washer and nut. way 3 way 4 way 5 way 6 way 8 way 9 way 10 way 12 way

 5 way 6 way 8 way 9 way 10 way 12 way

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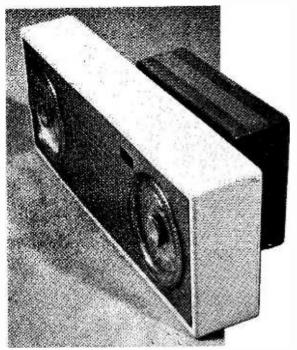
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How is it done? By picking up the Radio 2 signal on 200kHz (1500 metres) and mixing it with a local oscillator 429kHz in the Add-a-Verter thus producing a new signal on 629kHz which can be heard on the receiver on the medium wave band.

Cost of the bits and pieces bought new? About £2, plus the cost of the plastic box. The average constructor will be able to find most of the parts in his junk box.

CIRCUITRY

Nothing very complicated as can be seen in Fig. 1. A commonly available 2N2926 n.p.n. silicon transistor has a quartz crystal connected between its base and collector with the supply voltage being applied via

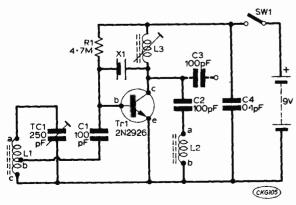


Fig. 1: The simple circuit of the Add-a-Verter. The output coil L2 is coupled to the ferrite rod aerial in the receiver, as shown in the heading photograph.



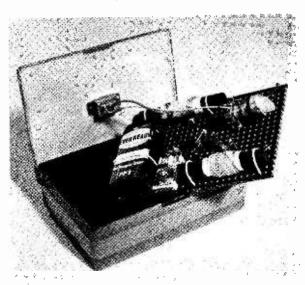
L3 which presents a high impedance at the crystal frequency. The circuit will still oscillate if the bottom end of C1 is connected to the negative line. But more later on the operation of the crystal.

The Radio 2 signal is picked up on the ferrite rod aerial L1 and fed to the base of the transistor. The signal at the collector now consists of the oscillator frequency 429kHz plus and minus 200kHz as well as the basic oscillator frequency of 429kHz. However it is the sum frequency of 629kHz which interests us, since it falls in the m.w. band.

The output coil L2 is also in the form of a ferrite rod aerial and it is the signal from this coil on 629kHz that is picked up by the ferrite rod aerial in the adjacent receiver.

THE CRYSTAL

The use of a crystal may seem extravagant and unnecessary to some but in fact it costs just 25p and its use saves much time fiddling around with coils and capacitors trying to hit the right frequency. Its stability is also so very much better than an LC combination. It is worth noting that the crystal oscillated immediately the prototype was switched on for the first time. The only other adjustment then needed



The only 'control' on the Add-a-Verter is the on-off switch!

was to peak trimmer TCl for maximum strength of the Radio 2 signal.

This particular type of crystal was originally used in equipment that multiplied its basic frequency fifty-four times so the crystal is designated as having an output frequency of 23·2 MHz which is shown on the crystal case as Channel 32. In fact its fundamental frequency is 429·630kHz. There is a whole range of similar crystals between about 370 and 540kHz. Some extremely useful tables showing the relationship between channel number, fundamental and output frequency can be found in the Radio Data Reference Book published by the RSGB.

To avoid interference with the converted Radio 2 signal from other stations on the medium wave band a search was made for a "quiet" spot at the l.f. end of the band but such a rarity disappeared a long time ago! The choice of the final frequency is therefore something of a compromise. What may seem a relatively quiet spot on the band in one part of the country may be alive with stations in another. However the Radio 2 signal appearing at the receiver was strong enough (in Surrey) to eliminate any other stations that can normally be heard at night on 629kHz. During the daytime there is no problem at all.

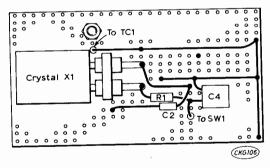
CONSTRUCTION

flat

2N2926

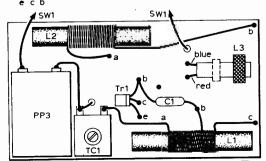
First obtain a plastic box, with lid, about 4×3 in. with a minimum depth inside of about 1^1 4in. The inside dimensions of the box will determine the size of the board on which the Add-a-Verter is built. The box used in the prototype was 4^1 4 $\times3\times1^1$ 4in. so the following description and accompanying diagrams are based on this.

The important factors to be observed, regardless of



▲ Fig. 2: Layout of the few components below the circuit board.

▼ Fig. 3: Component layout on top of the circuit board. Components are wired between Veroboard pins inserted in the board.



Board 4"x 234" but see text (CKG107)



Top view of circuit board, which may be compared with Fig. 3

the size of the board, are maximum separation between L1 and L2 while keeping them parallel and horizontal and keeping L3 away from both L1 and L2 to minimise interaction.

Cut a piece of plain Veroboard to the size required and insert Veroboard pins as shown in Fig. 2. The use of copper strip board was considered unnecessary and a waste with so few components to be mounted. Coils L1 and L2 are part of a m.w./l.w. ferrite rod aerial although any similar rod aerial can be used.

First remove the coils from the rod and then cut off two pieces of the rod 2^1 4in. long. If space permits simply cut the rod in half. The cutting is best done by nicking the rod at the appropriate place with the sharp edge of a file. Either hold the rod in a vice and tap the free end or strike the rod smartly on the sharp edge of a table or bench. If the resulting ends are a bit rough, don't worry, it won't make any difference!

The connections to the coils L1 and L2 are shown in Fig. 4. L2 should have the few turns constituting the base winding removed although this is not strictly necessary provided the ends of this winding are not allowed to short together. If difficulty is experienced in identifying the windings connect an ohmmeter to

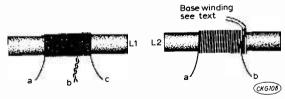


Fig. 4: Diagram of the lead-out wires of coils L1 and L2.

the single lead "a" and find the other end of the winding "b". The remaining ends will be the base winding.

Put the coils on the ferrite rods and attach the rods to the board with loops of thin insulated wire. Drill a hole and fix the compression trimmer TC1 to the board with the lock nut provided. Place the tags of the crystal holder over the appropriate pins on the underside of the board and solder, see Fig. 2. These pins provide anchoring points for the base and collector leads of the transistor on top of the board.

Coil L3 is the long wave winding of a l.w./m.w.

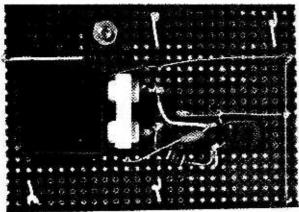
pins

coil, freely available, and intended for use in a simple "crystal set" circuit. An iron core to suit the coil is required and should be bought at the same time as the coil since the core is not normally sold as part of the coil. In the prototype the red and blue tags of the l.w. winding were soldered directly to the appropriate pins in the board, Fig. 3, but if height permits mount the coil normally by its fixing lugs. Initially the core should be screwed in as far as possible.

Rather than cut off the excess of the coil lead-outs it is advisable to use the tinned ends as supplied, winding any surplus wire back on to the coil formers.

Complete the wiring in of the components as shown in Figs. 2 and 3 leaving the transistor until last. Hold the leads of the transistor in a pair of flat nosed pliers while tinning the ends and soldering them to their pins. Transistors are pretty tough these days but there is no point in tempting Fate!

Finally, attach the PP3 battery to the board with a wire loop as was done with the ferrite rods. The negative side of the battery clip goes to the general negative line while the positive side goes to the on-off switch, mounted on the lid of the box, and back to the positive side of the circuit. The current taken by the prototype was only 120µA so there was a temptation to omit the switch altogether! In this case the



Photograph of the underneath of the circuit board with connections as shown in Fig. 2.

working life of the battery would be about the same as its shelf life.

Plug in the crystal which should lie flat against the board but if necessary fix it to the board with another loop of wire.

SETTING UP

Put the board into the case where it should lie flat on the case of the crystal. No additional fixing was thought to be necessary. Connect a milliammeter in series with the positive battery lead. If using a multimeter set it to the maximum current range in case there are any short circuits! Switch on and if the crystal oscillates first time the current will be of the order of $120\mu A$. If not the current will be about $200\mu A$ when the core of coil L3 should be unscrewed slowly the current dropping to a minimum and rising again. As a final check on the crystal, switch the unit on and off a few times. Each time the current should fall to around $120\mu A$ but if not readjust the core until it does. To be quite sure, remove the crystal, when the current should rise.

Switch on the portable radio and set the dial to

* components list

R1 4 7M2 tw 100oF CI C2 100pF C3 100pF (if required) C4 0-14F 12VW TC1 250pF compression trimmer Tr1 2N2926 (grange, yellow or green coding) L1/L2 part of territe rod aerial (Type MW/LW/5FR Denco) part of I.w./m.w. coil, with core (Type DRX1 Henrys) crystal type FT241A, approx. 429kHz ("Channel 32-23-2MC", Henrys) and holder Miniature on off switch, PP3.9V battery and clips. Veroboard 0-15in matrix or plain paxolin, 4 × 21in (see text). Plastic box.

around 480 metres on the medium wave band. Stand the Add-a-Verter close behind the radio so that the output coil is close to the ferrite rod aerial in the set and parallel to it. There is a fair chance that Radio 2 will now be audible but if not tune around 480m. on the set at the same time adjusting trimmer TC1 in the Add-a-Verter until Radio 2 is heard. Adjust the trimmer for maximum volume. To check, move the Add-a-Verter away from the set when the Radio 2 signal should disappear. For portability the Add-a-Verter can be attached to the radio with a couple of rubber bands.

Should the m.w. receiver not have ferrite rod aerial the Add-a-Verter can still be used by adding capacitor C3 from Trl to a terminal on the box and joining the terminal to the aerial socket on the receiver.

NOTES

Do not use a metal box since operation of the Add-a-Verter depends upon the coupling between coil L2 and the ferrite rod aerial in the receiver. A metal box will act as a screen preventing reception of Radio 2. A metal box is only permissible if L2 is mounted outside and at least 12 in. clear of the metal box, but this is rather getting away from the original idea of the Add-a-Verter.

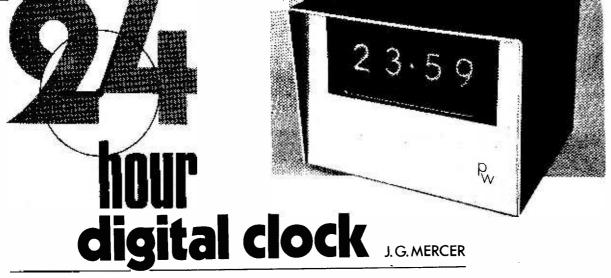
Don't forget that ferrite rod aerials are highly directional and that the set and Add-a-Verter must be orientated for maximum signal strength of Radio 2.

Switch off the Add-a-Verter when normal reception is required on the medium waves as otherwise interfering beat notes will be heard on stations.

Transistors type 2N2926 with orange, yellow and green coding were all tried and found satisfactory but difficulty may be experienced if the lower gain types, brown or red, are used.

BACK NUMBERS

We regret that the back numbers department of P.W. has now closed and consequently we are unable to supply these Requests for specific back issues can usually be included in our 'CQ' section; there is no charge for this but it is a service between readers and P.W. are unable to meet any of these requests.



PART 2

NOTE:—In the Components List in Part 1 of the 24 Hour Digital Clock the reference number of the monolithic voltage regulator should have been shown as MVR5V and not MRV5V. The resistor marked R3 in Fig. 1 should have been shown as R4.

CONSTRUCTION

Construction commences with the assembly of four decade counter boards for the display module. First of all four holes should be drilled at the extreme corners of the boards in order to take the 6BA studding with which the display module is finally assembled, plus another below the edge connector, Fig. 7.

DC 3 DC 4 DC 2 To SW1/W DC.1 (Decade counter board 1) Tens minutes Units hours Tens hours Units minutes Numicator tube 0 LINK ♠ A O/P OD O/P O CLOCK INPLIT **◆**ZERO RESET O O/P NEXT DECAD NINE RESET GROUND (B O/P 0 OC O/P 0 0 0 0 0 0 0 0 O DEC. POINT RIGHT O DEC. POINT LEFT 0 0 To countdown To countdown To power module (M 3)♥ module (M1) tagstrip (D)

Each board has the component values and locations clearly marked together with the relevant notation for the edge connections. The resistor marked R1 on each of the P.C. boards corresponds to the R1's on the circuit diagram and has a value of $16k\Omega$ for the recommended tube. Tinned copper wire should be used for all the links shown on the boards. The two types of integrated circuits should now be mounted on the boards paying especial attention that the indentations on the i.c.'s align with the marks on the P.C. board.

The numicator tube is now mounted with its leads straddling the board and viewing direction to the front. The pin connections for the tube are shown in Fig. 3 and the positions for these connections are clearly shown on the board itself.

All four boards are assembled in exactly the same way. Three of the four boards must now be slightly modified as shown in Fig. 6 (boards 2, 3 and 4). This modification is made to separate the two zero reset connections in order to be able to modify the counting sequence as described in the circuit operation.

The boards are now assembled on the studding as shown in Fig. 7 and spaced using extra 6BA

Fig. 7. Assembly and interconnecting wiring of the display module comprising the four decade counter boards.

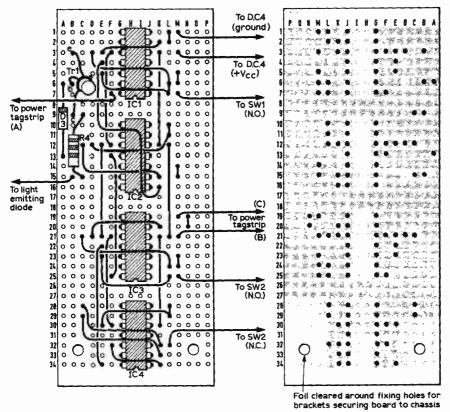
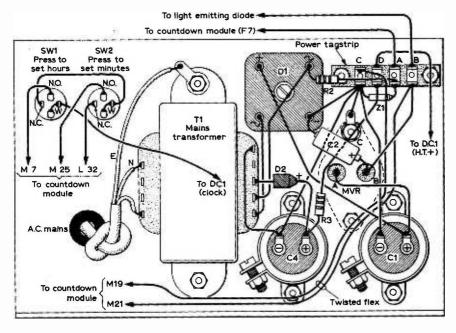


Fig. 8. Constructional details and wiring of the countdown board. Care must be taken to ensure that the copper strips are cleared from around the two fixing down brackets.

▼ Fig. 9. Layout of the power supply components and two reset switches on the back panel. The voltage regulator must be bolted directly on the back panel which acts as a heat sink.



nuts. Mounting feet are fitted to the P.C. boards—these can be seen in the photographs.

The interconnections at the rear of the boards are now made as shown in Fig. 7. Connections may be made with tinned copper wire through the holes in the edge connectors. Flying leads are connected for the input, +5V, chassis, and h.t. voltage. This completes the display module.

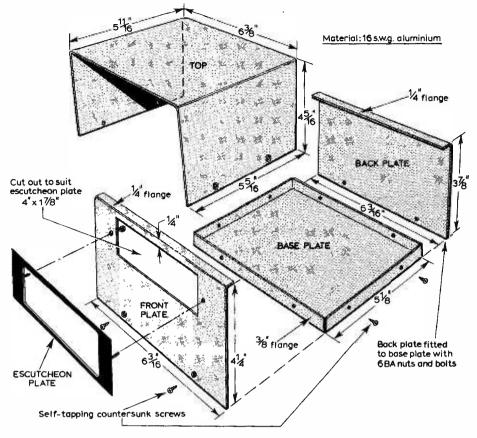
The countdown board is now assembled on a piece of 5 lin. matrix stripboard 334in. long by 114in. wide. The layout is exactly as shown in the circuit diagram, the only cuts to be made to the board are two complete lines along the length of the board between the rows of i.c. pins, Fig. 8.

All the chassis connections of the i.c.'s are commoned as are the 5V pins and flying leads added for connection to the power supply. Flying leads are also connected for the input, output and setting switches. This completes the countdown module.

The main chassis on which the power supply and other modules are mounted is simply an aluminium panel (18 s.w.g.) with an L cross section. The rear panel carries the setting switches and power supply while other modules are bolted to the base. The transformer is bolted to the rear panel as is the monolithic voltage regulator which uses the rear panel as its heat sink, Fig. 9.

The two large capacitors are also fixed to the rear panel with vertical fixing clips. The remaining power supply components may be soldered to the tags of the components already mentioned and the four outlets taken to a tag panel to permit easy connection to the other two modules.

The countdown board and display module are now bolted on to the basic chassis using angle



Dimensions and drilling details for the case and chassis which can be obtained as a kit. Note that the flanges on the base plate are uppermost. The backplate is $\frac{3}{6}$ in shorter than the front plate thus leaving a slot along the back to provide essential ventilation. The metalwork shown in the various photographs of the Digital Clock was that used by the author in the prototype.

brackets and all the flying leads connected to the power supply and setting switches on the rear panel.

SETTING UP

The initial setting of the clock to a particular time is carried out as follows:

SW1 is depressed to allow the clock to run at PRESENT SETTING SYSTEM

SW₁ SW2 Display Press to set Press to set Output hours minutes from IC4 ¹∕₃Hz output 25Hz output from IC3 from IC1 NEW SETTING SYSTEM 25Hz output 1/2Hz output from IC1 from IC3 SW1 SW2 Press Press to set to set hours Each gate \bigcirc is $\frac{1}{4}$ 7400 I.C. \therefore 3 7400 I.C.'s are required Fig. 10. Output from IC4

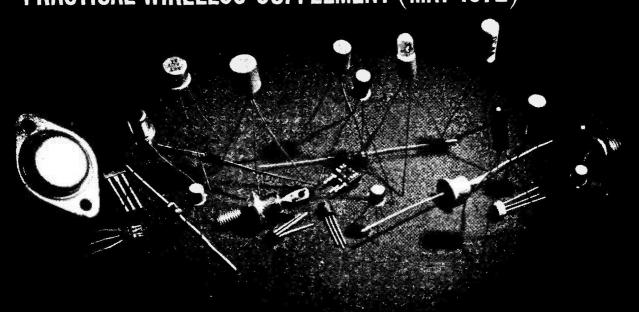
1500 times normal speed and released at a displayed time at least half an hour before the required setting. SW2 is now depressed which allows the clock to run at a speed 30 times greater than normal. The button is released at the desired time setting. The clock will now run to an accuracy determined only by the accuracy at which the mains frequency is maintained. Resetting after a power cut or acci-

dental disconnection may be carried out as described above.

In some cases and with some types of switches, contact bounce can cause overshoot on the initial setting. It is possible to use extra integrated circuits together with changeover switches to provide electrical rather mechanical control of the initial setting. However, due to the fact that the clock will only infrequently need setting, the extra circuit complexity and expense was not considered justified. A circuit overcome contact bounce is shown in Fig. 10.



semiconductor data practical wireless supplement (MAY 1972)



COMPILED FOR PRACTICAL WIRELESS BY D.A. LONGLAND, B.Sc., C. Eng., M.I.E.R.E. (ELECTROVALUE)

THIS SUPPLEMENT has been designed to present data to readers in a basic way. The technical details, parameters and capabilities of the more common types are given. The tables are broken down into categories by usage to simplify selection but of course many transistors could be included in several different headings. This has not been done as it would necessitate a considerable degree of duplication.

Transistors are used for many purposes, most of which may be reduced into the broad classification of increasing the power or level of a signal. This is effected by introducing the signal into the base-emitter (or gate-source circuit if a field effect transistor, f.e.t.) and thus causing an amplified variation of the signal in the collectoremitter circuit (or source drain circuit if an f.e.t.).

The important characteristics required of a transistor are determined by its application, that is, whether used for amplification of small a.c. or d.c. signals, for switching current or power amplification. Other characteristics may be important when a transistor is used near one of its limits, for example, high frequency, low noise, saturation, high power, very low current, or fast switching times.

Transistor characteristics

An ideal transistor will have a high transition frequency (fT) and low feedback capacitance (for good high frequency response), stable characteristics, low noise level, low saturation voltage, fast rise and fall times (for low loss switching characteristics), high thermal conductivity (for high power handling capacity), low leakage (for ability to operate at very low currents), high gain (to reduce the number of stages for a given requirement) and low temperature effects (to reduce dependence on operating temperature and power dissipation).

It would be rare to find many extreme characteristics in one transistor for the technology used in producing an extreme in one characteristic may be incompatible with that used in producing an extreme in another. Hence, if a small signal, high frequency amplifier transistor were chosen it would have a medium gain at low currents, low feedback capacitance and high fT, therefore it would be made on a small semiconductor chip. One would not therefore expect the same transistor to handle a high voltage, high power, or have a low saturation voltage at high current. This illustrates the need for com-

promise and in turn accounts for the large number of transistors available, many having their own individual collections of characteristics suited to particular applications.

In choosing a transistor, one has to start by determining the category in which the major characteristics fall, e.g. r.f. or a.f. amplification, switching, low or high power, high or low voltage, PNP or NPN. After choosing a range of transistors, the secondary characteristics may then be chosen.

- 1. In high frequency application, having chosen one with an fT considerably higher than the operating frequency (if possible) and having a low feedback capacitance one would then seek a type having suitable voltage rating, power, dissipation, current or power gain, and if for switching, rise and fall time and saturation voltage.
- 2. For low frequency application, the fT is relatively unimportant though with a high gain transistor, a transition frequency rather higher than the product of the current gain and highest frequency of operation should be sought: thus an amplifier using a transistor with an fT of 1MHz and a current gain of 100 would be expected to fall in frequency response at 10kHz. A suitable voltage and power rating would then be chosen, keeping in mind any extremely low current or low saturation voltage characteristic required: thus an amplifier to operate at 1mA should have a collector-base leakage current rather less than 1mA divided by the desired maximum current gain of the transistor in order that the transistor will not saturate. A high gain transistor will give a higher input resistance than a low gain type, thus the high gain type would take less to drive power, but may cause a limitation in the frequency response.
- 3. For switching application it is desirable that the transistor conduct heavily with little voltage drop (i.e. low saturation voltage) when driven. For high speed switching the response time (i.e. rise, fall and storage) is important since the power dissipated in a slow transistor will be relatively high and may limit the power or frequency to be switched.
- 4. Other characteristics of a particular nature

may be sought, e.g. a given outline for incorporation into existing equipment, or ease of heat sinking (to reduce temperature rise if high power is to be dissipated), or high reverse base voltage for certain switching purposes, etc., etc. All must be considered in the design of equipment and in the selection of a suitable or substitute transistor.

The following tables are designed to assist in selection and substitution—it will be found that perhaps many transistors will serve one particular and non-demanding application whereas for another application only one or two will be found suitable. One factor to be considered is cost as it will be observed that a transistor which incorporates several special or extreme characteristics can be expensive. Good, common transistors which are relatively cheap and which will fulfil many needs are indicated by a dagger.

Dissipation in power transistors

Under the High Current Audio headings, the power dissipation is given in terms of thermal conductivity between junction and case in mW/°C. There is some difficulty in expressing dissipation concisely as manufacturers are inclined to give dissipation ratings at different temperatures. It is felt that the thermal conductivity gives a more useful guide in the comparison between transistors.

In order to calculate the dissipation at a particular case temperature one must multiply the excess of maximum junction temperature above case temperature by the thermal conductivity. In some cases the manufacturer does not declare the dissipation at 25°C, therefore one should not assume that it would be safe to dissipate the maximum power as calculated above at 25°C case temperature. One may assume it safe to operate at a case temperature midway between 25°C and maximum junction temperature, this covering most practical requirements.

As a general guide, the maximum junction temperatures (which should not be exceeded by 20°C) are: Germanium (metal case) 90°C: Silicon (metal case) 200°C; Silicon (plastic case) 150°C for types 2N-, MJE-, TIP- and 125°C for types BD-.

Ahhre	eviations	lb	Base current.
ADDIC		lc	Collector current.
†	Inexpensive types	Icbo	Collector-base current with emitter
t	Typical.		open-circuit.
max	Maximum.	NF	Noise factor.
min	Minimum.	PG	Power gain.
Cob	Maximum common-base output capac- itance.	Vcbo	Collector-base voltage with emitter open-circuit.
f	fhfb, Frequency at which the common-	Vce	Collector-emitter voltage.
	base current is reduced by 3dB.	Vce(sat)	Collector-emitter voltage for saturated
fT	Frequency at which the common-	,	(fully conducting) operation.
	emitter current gain is unity.	Veb	Base-emitter reverse voltage.
hFE	Large signal current gain.	*	Under Ic/Ib. These refer to Ic with the
hfe	Small signal current gain.		collector and base shorted.

GERMANIUM TRANSISTORS

R.F. Applications, PNP

Type	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hrg *h _{fe}	l _e mA	l _{cbo} μ A	Vc• (sat)	I _c /I _b mA	f _T MHz	С _{ов} pF	NF dB max	PG dB	MHz @	Base
NKT613F	40	40	1	10	80	*40 min	1	5			75t	2		14	100	TO7
AF114	20	20		10	60	*130t	1	8	1		75t	2.5	1	12.5	100	TO7
AF115	20	20	1	10	60	*130t	1	8)		75t			10	100	TO7
AF116	20	20	1	10	60	*130t	1	8	1		75t		ļ	19	10.7	TO7
AF117	20	20		10	60	*130t	1	8	1		75t			40	0.45	T07
AF118	70	70	1	30	375	*180t	1	6	į		175	2.3		1		TO7
AF124	20	20		10	60	*130t	1	8			75t	2.5		12.5	100	TO72b
AF125	20	20	1	10	60	*130t	1	8	1		75t			10	100	TO72b
AF126	20	20	ļ	10	60	*130t	1	8	i i		75t			19	10.7	TO72b
AF127†	20	20	ļ	10	60	*130t	1	8	1		75t			40	0.45	TO72b
AF139	20	15	0.3	10	60	50t	1	8			500t			10	800	TO72a
AF201	25		0.3	10	225	50t	1	10	1						1	TO72b
AF239†	20	15	0.3	10	60	50t	1	8			700			14	800	TO72a
OC44	15		12	10	75	40-225	2	2			7·5f			l		TO1,RO
OC45	15	1	12	10	75	25-125	2	2			9f			l	1	TO1, RO

Low Noise, PNP

Туре	V _{cbo} max	V _c . max	V _{eb} max	I _c mA max	P _{tot} max	hFE *hfe	I _c mA	I _{cbo} μΑ	V _{ce} (sat)	Ic/Ib mA	f _T MHz	C _{ob} pF	NF dB max	PG MHz dB	Base
AC107 AC126 AC151R†	15 32 32	15 12 24	5 10 10	10 100 200	80 220 150	60t *130–300 80t	0·3 2 2	13 10 100			0·3f 2·3 1·5	14 27	5 10		T01, R09 T01 T01

Small Signal Audio, PNP

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	i _c mA max	P _{tot} max	hfE *hte	I _c mA	lebo μΑ	V _{ce} (sat)	l _e /l _b mA	f _T MHz	C _{ob} pF	NF dB max	PG MHz dB @	Base
NKT213 NKT214† NKT274† NKT275 OC71	32 32 15 15 30	32 32 15 15 20	10 10 5 5	250 250 250 250 250 10	200 200 200 200 200 125	50-130 *30-75 *85-250 *30-90 30-75	1 1 1 1 3	10 10 10 10 10	0·2 0·2	25/2·5 25/2·5	0·9f 0·9f 1 1 0·3f	60 60	10 15 10 16		TO1 TO1 TO1 TO1 TO1,RO9

Medium Current Audio, Small Output Types, PNP

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hpg *hfe	I _c mA	I _{cbo} μ A	V _{ce} (sat)	lc/lb mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
AC128† AC142H AC142HK AC153 AC153K† NKT211† NKT212 NKT217 NKT271 OC72 OC81 OC83 OC84†	32 50 50 32 32 32 32 60 15 32 32 32 32	16 30 30 18 18 30 32 40 15 32 32 20	10 10 10 10 10 10 10 10 5 10 10 3	1000 1200 1200 2000 2000 1000 500 500 500 500 500 500	220 220 860 220 1000 200 200 200 200 125 240 220 220	40-110 40-110 40-110 50-250 50-250 20-150 50-150 50-150 50-250 30-90 50-250 50-280 60-200	300 400 400 300 300 300 50 25 50 80 50 50	10 14 14 200 200 10 10 10 10 15 10 100 100	0·5 0·15 0·5 0·25	300 /30 25/2·5 300/6 125/12·5 300/9	1·5 1·2 1·2 1·5 1·5 0·9f 0·9f 1 1 1 1 0·85 0·85	100 60 60				T01 T01 X9a T01 T01 T01 T01 T01, R00 T01 T01 T01

GERMANIUM TRANSISTORS CONTINUED

Medium Current Audio, Small Output Types, NPN

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hFE *hfe	I _c mA	I _{cbo} μΑ	V _{ce} (sat)	I _c /I _b m A	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz	Base
AC127† AC141H AC141HK AC176 AC176K† NKT713 NKT773	32 50 50 32 32 30 15	32 25 25 20 20 30 15	10 10 10 5 5 15	500 1200 1200 300 300 500 300	200 220 860 220 1000 150	25-143 40-110 40-110 52-180 52-180 50-150 50-150	500 400 400 500 500 50 50	10 14 14 100 100 15 15	1 0·55 0·55 0·6 0·6 0·15	500/50 400* 400* 500/10 500/10 50/5	1·5 3 3 1 1 1 1f					TO1 TO1 X9a TO1 X9a TO1 TO1

High Current Audio, Power Output Types, PNP

Туре	V _{cbo} max	V _{ce} .	V _{eb} max	I _c max	P _{tot} max	h _{FE} *h _{fe}	I _c mA	I _{cbo} μΑ	V _{ce} (sat)	1 _c / I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
AD140 AD142 AD149† AD162† AL102 NKT403† NKT405 OC25 OC28 OC29† OC35 OC36 2N2147	55 80 50 32 130 80 60 40 80 60 60 80 75	50 30 20 60 32 32 40 60 48 48 60 50	10 10 20 10 2 40 20 10 40 20 40 20 40 1.5	3A 10A 3·5 A 1A 6A 10A 5A 4A 8A 8A 8A 8A	666 666 500 222 666 770 770 500 666 666 666 666 666	30-100 30-170 30-170 80-320 40-250 50-150 100-200 15-80 20-55 45-130 25-75 30-110	1000 1000 1000 500 1000 1000 1000 1000	100 5000 350 40 1000 150 100 100 100 100 100	0·3 0·5 0·42 0·42 1·6 1·6 1·4 1·6 0·6	5000/250 5000/250 1000/100 1000/100 6000* 6000* 6000* 5000/250	0·45 0·5 1 4 0·35 0·35 0·25 f 0·25 f 0·25 f 0·25 f					TO3 TO3 TO3 MD17c TO3 TO3 TO3 TO3 TO3 TO3 TO3 TO3 TO3 TO3

High Current Audio, Power Output Types, NPN

ĺ	AD161+	32	20	10	1A	222	80–320	500	50		1			MD17c	
	ADIOIT	32	20	10	'^	222	80-320	300	30		•	ļ			ı

See note on power dissipation

Medium Current Switching, PNP

	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hre *hre	I _c mA	Ι _{εδο} μ Α	V _{ce} (sat)	I _c /I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
ACY17 ACY18 ACY19 ACY20 ACY21 ACY21 ACY22 ACY39 2N1303 2N1305† 2N1307 2N1309 ASY26 ASY27	70 50 50 40 40 20 110 30 30 30 30 25	32 30 30 20 20 15 40 25 20 15 15 15	12 12 12 12 12 12 12 12 25 25 25 25 20 20	500 500 500 500 500 500 500 200 200 200	260 260 260 260 260 260 260 150 150 150 150	50-150 40-120 80-250 50-145 90-250 30-300 50-150 20 min 40-200 60-300 80 min 30-80 50-150	300 300 300 50 50 300 300 10 10 10 20 20	10 10 10 10 10 10 10 10 6 6 6 7	0·3 0·3 0·2 0·2 0·3 0·3 0·2 0·2 0·2 0·2 0·25	300/15 300/15 300/15 50/1·3 50/1·3 300/15 300/15 10/0·5 10/0·5 10/0·17 10/0·13 50/2 50/1·55	1 1 1·5 1 0·8 1 5 10 15 20 8 14	20 20 20 20 20 15 16				TO5 TO5 TO5 TO5 TO5 TO5 TO5 TO5 TO5 TO5

Medium Current Switching, NPN

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hfe *hfe	I _c mA	l _{cbo} μΑ	V _{ce} (sat)	l _c /l _b mA	f _T MHz	C _{ob} pF	NF dB max	PG MHz dB	Base
ASY28 ASY29 2N1302 2N1304† 2N1306 2N1308	30 25 25 25 25 25 25	15 15 25 20 15 15	20 20 25 25 25 25 25	200 200 200 200 200 200 200	150 150 150 150 150 150	30–80 50–150 20 min. 40–200 60–300 80 min.	20 20 10 10 10	35 35 6 6 6	0·25 0·25 0·2 0·2 0·2 0·2 0·2	50/2 50/1 · 25 10/0 · 5 10/0 · 25 10/0 · 17 10/0 · 13	14 20 10 15 20 30	16 16 20 20 20 20			TO5 TO5 TO5 TO5 TO5 TO5

High Current Switching, PNP

Туре	V _{cbo} max	V _{eb} max	I _c max	P _{tot} max	hre *h _{fe}	I _c mA	lebo μA	V _{ce} (sat)	l _c /l _b mA		C _{ob} pF	NF dB max	PG dB	MHz	Base
AU111	320	2	10A		15-80	6000		0.5	5000/250	2					ТО3

The term switching above does not imply any discontinuity in the characteristics. Such transistors may be used as amplifiers, as may amplifier transistors be used for switching duty if their characteristics are suitable.

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R.F. Applications, NPN

*ARRETTY OF

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hre *h _{fe}	I _c mA	l _{cbo} μΑ	V _{ce} (sat)	I _c / I _b mA	f T MHz	C _{ob} pF	NF dB max	PG dB	MHz a)	Base
BF115	50	30	5	30	145	100t	3				230t		4			TO72b
BF167	40	30	4	25	130	50t	4				350t		3.5			TO72b
BF173†	40	25	4	25	260	87t	7		1	1	550t			1		TO72b
BF194†	30	20	5	30	220	115t	1		!!	1	260t		4)	1	MM10b
BF195	30	20	5	30	220	67t	1		!!!		200t		4)	1	MM10b
BF254†	30	20	5	30	220	115t	1		1	1	260t		À			TO92z
BF255	30	20	5	30	220	67t	1		1 1	1	200t		4	1	l	·TO92z
BFY90	30	15	2.5	50	200	60t	2		1 1	i	1000	1.8	5	22t	500	TO72a
2N3663	30	12	3	25	200	75t	8	0.5	1		1250t	1.7	6.5	15	200	TO98
2N4292†	30	15	3	500	200	20 min	3	0.5	i i		570t	2.2	6	29t	100	U29

Small Signal Audio, PNP

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hfE *h _{fe}	l _c mA	l _{cbo} μ A	V _{ce} (sat)	t _c /i _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
BC157†	50	45	6	100	220	*125-250	2	0.05	0·2t	100/5	130t					MM10a
BC158	30	20	5	100	220	*125-500	2	0.1	0·2t	100/5	130t		1		1	MM10a
BC177†	45	45	5	100	300	*125-260	2	0.05	0·2t	100/5	130t	4	1	ł		TO18
BC178	20	20	5	100	300	*125~500	2	0.1	0·2t	100/5	130t	4	1	Ì		TQ18
BC186	40	25	5	100	300	40 min	2				60	5		į	1	TQ18
BC212L	60	50	5	100	360	60-300	2	0.015		10/0 - 5	200	10	Í		1 .	TQ92a
BC213L	45	30	5	100	360	80-400	2	0.015		10/0.5	200	10	İ	1	1	TQ92a
BC257†	50	45	5	100	220	*125-260	2	0.1	0.5	100/5	130t	6	i			TQ92a
BC258	30	25	5	100	220	*125-500	2	0.1	0.5	100/5	130t	6	1		1	TQ92a
BCY70	50	40	5	200	350	50 min.	10	0.01	0.25	10/1	250		ļ į		1 1	TO18
BCY72	25	25	5	200	350	50 min.	10	0.1	0.25	10/1	200		!	1	1	TO18
ZTX500	25	25	5	500	300	50-300	10	0.2	0.35	50/5	150	6	· ·			X59
ZTX501	35	35	5	500	300	50-300	10	0.2	0.25	50/5	150	6	i '			X59
ZTX502	35	35	5	500	300	100-300	10	0.2	0.25	50/5	150	6				X59
ZTX503	45	45	5	500	300	50-300	10	0.2	0.35	50/5	150	6			1 1	X59
ZTX504	70	70	5	500	300	50-300	10	0.2	0.35	50/5	150	6	[1 1	X59
ZTX530	30	30	5	500	250	100-400	0.1	0.2	0.7	10/0 · 5	30	8			1 1	X59
ZTX531	45	45	5	500	250	40-120	0.01	0.2	0.7	10/0-5	30	8			1 1	X59
2N3906	40	40	5	200	310	100-300	10	0.05	0.25	10/1	250	4.5			1 1	TO92b
2N4126	25	25	4	200	310	120-360	2	0.05	0.4	50/5	250	4.5			1 1	TO92b
2N4059	30	30	6	30	250	45~660	1		0.7	10/0·5					1 1	TQ92a
2N4060	30	30	6	30	250	45165	1		0.7	10/0-5					1 1	TO92a
2N4061	30	30	6	30	250	90-330	1		0.7	10/0 - 5						TO92a
2N4062	30	30	6	30	250	180-660	1		0.7	10/0 - 5					1 1	TO92a
2N4289	60	45	7	500	250	160 min.	1	0.01	0.35	1/0-1	170t	8			1 1	U29

SILICON TRANSISTORS CONTINUED

Small Signal, Audio, NPN

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hre *hfe	I _c mA	lebo μ A	V _{ce} (sat)	I _c /I _b mA	f _T MHz	С _{ов} pF	NF dB max	PG dB	MHz ā	Base
BC107†	45	45	5	100	300	*125-500	2	0.01	0.25	10/0-25	150	3.7	10			TO18
BC108	20	20	5	100	300	*125-900	2	0.01	0.25	10/0.5	150	3.7	10		Į	TO18
BC147†	50	45	6	200	220	*125-500	2	0.02	0.25	10/0.5	300t	2.5	10		1	MM10a
BC148	30	20	5	100	300	*125-900	2	0.25	0.25	10/0.5	300t	2.5	10		i	MM10a
BC167†	50	45	6	100	300	*125-500	2	0.2	0.25	10/0.5	300t		10	1	ļ	TO92a
BC168	30	20	5	100	300	*125-900	2	0.2	0.25	10/0.5	300t		10	ł	1	TO92a
BC182L	60	50	5	200	300	100-480	2	0.015		10/0-3	150	5	10	1	ĺ	TO92a
BC183L	45	30	5	200	300	100-400	2	0.015			150	5	1	į]	TO92a
BC267	50	45	6	1000	375	*125-500	2	0.015		200/10	150	١	3t	1	i	TO18
BC268	30	20	6	1000	375	*125-900	2	0.015		200/10	150		3t			TO18
ZTX300	25	25	5	500	300	50-300	10	0.013	0.35	50/5	150	6	31	1	1	X59
ZTX301	35	35	5	500	300	50-300	10	0.2	0.25	50/5	150	6	1	İ]	X59
ZTX302	35	35	5	500	300	100-300	10	0.2	0.25	50/5	150	6		1	1	X59
ZTX303	45	45	5	500	300	50-300	10	0.2	0.35	50/5	150	6	1	1		X59
ZTX304	70	70	5	500	300	50-300	10	0.2	0.35	50/5	150	6	i .			X59
ZTX330	30	30	5	500	250	100-400	10	0.2	0.7	10/0.5	30	8				X59
ZTX331	45	45	5	500	250	40-120	10	0.2	0.7	10/0.5	30	8	ľ	1		X59
2N2924	25	25	5	100	200	*150-300	2	0.5	٠. ا	10,00	120t	12	2 · 5t	i	l	TO98
2N2925	25	25	5	100	200	*235-470	2	0.5			120t	12	2.5t	ĺ		TO98
2N2926B	18	18	5	100	200	*35-70	2	0.5	į		120t	12	2·5t			TO98
2N2926R	18	18	5	100	200	*55-110	2	0.5			120t	12	2·5t	ļ	ĺ	TO98
2N2926O	18	18	5	100	200	*90-180	2	0.5			120t	12	2.5t	ì	1	TO98
2N2926Y†	18	18	5	100	200	*150-300	2	0.5			120t	12	2·5t			TO98
2N2926G†	18	18	5	100	200	*235-470	2	0.5			120t	12	2.5t	İ	İ	TO98
2N3708	30	30	6	30	250	45-660	1	0.1	1	10/0.5	,			ł	Ì	TO92a
2N3709	30	30	6	30	250	45-165	1 1	0.1	1	10/0-5			l		l	TO92a
2N3710	30	30	6	30	250	90-330	i i	0.1	1	10/0.5	· '			1		TO92a
2N3711	30	30	6	30	250	180660	1	0.1	1	10/0.5	1			1	1	TO92a
2N3904	60	40	5	100	310	100-300	10	0.1	0.2	10/1	200t	4				TQ92b
2N4124	30	25	5	200	310	120-360	2	0.05	0.3	50/5	300	4	1		!	TO92b
2N4286†	30	25	6	500	250	100 min	0.1	0.05		•	280t	6		1	i	U29
2N5172	25	25	5	100	200	*100-750	10	0.1	0.25	·10/1	120t	10	1	1	1	TO98

Low Noise, NPN

Type	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hfe *hfe	l _e mA	l _{cbo} μ A	Vce (sat) @	I _c /I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG N dB	IHz Base
BC109B†	20	20	5	100	300	*240-500	2	0.015	0.25	10/0.25	300t		4		TO18
BC109Ct	20	20	5	100	300	*450-900	2	0-015	0.25	10/0-25	300t		4	1 1	TO18
BC149†	30	30	5	200	220	*240-900	2	0.02	0.25	10/0 25	300t		4		MM10a
BC169†	30	30	5	100	300	*240-900	2	0.02	0.25	10/0-25	300t		4	l -1	TO92a
BC184L	45	30	5	200	300	250min	2	0.015		•	150		4		TO92a
BC269	30	20	6	1000	375	*240-900	2	0.015	0.22 t	200/10	150		4	1 1	TO18
2N930†	45	45	5	60	300	100-300	0.01	0.01		-	80t	8	3		TO18
2N2483	60	60	6	50	360	*40-120	0.01	0.01			60	6	4		TO18
2N2484	60	60	6	50	350	*100-500	0.01	0.01			60	6	3		TO18
2N3707	30	30	6	30	250	*100-400	0.1	0.1	1				5	}	TO92a
2N5088	35	30	3	50	310	*300-900	0.1	0.05			50	4	3	l í	TO92b

Low Noise, PNP

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hrg *hfe	I _c mA	lcbo μA	Vc. (sat)	I _c /I _b mA	fy MHz	C _{ob} pF	NF dB max	PG MHz dB	Base
BC153	40	40	5	100	200	50-135	0.1	0.002		10/0-5	40	6	2.5		RO97a
BC154†	40	40	5	100	200	160-330	0.1	0.002		10/0 5	40	6	2.5	! !	RO97a
BC159†	30	20	5	100	200	*125-500	2	0.05	0-3t	100/5	200t	5	4	1 1	MM10a
BC179†	20	20	5	100	300	*125-900	2	0.05	0.3t	100/5	200t	4	4		TO18
BC214L	45	30	5	100	360	140-400	2	0.015	0.2	10/0-5	200	10	2	1 1	TO92a
BC259†	25	20	5	100	220	120-460	2	0.1	0.8	10/0 - 5	130	6	4	i i	TO92a
BCY71	45	45	5	200	350	100 min	10	0.01	0.25	10/1	300		ļ		TO18
2N4058	30	30	6	30	250	100-400	0.1	0.1	0.7	10/1			5		TO92a

Medium Current Audio, Small Output Types, NPN

Туре	V _{cbo} max	V _c	V _{eb} max	I _c mA max	P _{tot} max	hfg *hfe	I _c mA @	I _{cbo} μ A	V _{ce} (sat)	I _c /I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
BC125	50	30	5		300	30 min	150	0.05	0.5	150/15	40	12				RO97
BC140-10	80	40	7	1000	750	20 min	1000	0.1	1 1	1A/100	50	25		İ	1	TO5
BC301	90	60	7	1000	850	40-240	150	0.02	0.5	150/15	120t	-			1	TO5
BFX84†	100	60	6	1000	800	30 min	150	0.05	0.35	150/15	50		į	i		TO5
BFX85	100	60	6	1000	800	70 min	150	0.05	0.35	150/15	50			l		TO5
BFY50	80	35	6	1000	800	30 min	150	0.05	0.2	150/15	60			Į.		TO5
BFY51†	60	30	6	1000	800	40 min	150	0.05	0.35	150/15	50		ļ			TO5
BFY52	40	40	- 6	1000	800	60 min	150	0.5	0.35	150/15	150				1	TO5
MC140	80	40	7	1000	800	40-300	150	0.5	0.7t	1A/100	100					X58b
2N696	60	40	5	500	600	20-60	150	1	1.5	150/15	40	35		ŀ	ļ	TO5
2N697	60	40	5	500	600	40-120	150	1	1.5	150/15	50	35	Ì		1	TO5
2N1613	75		7	500	800	40-120	150	0.01	1.5	150/15	60	25	İ	1	1	TO5
2N1711	75	30	7	1000	800	100-300	150	0.01	1.5	150/15	70	25			1	TO5
2N2270	60	45	7	1000	1000	50-200	150	0.05	0.9	150/15	60				1	TO5
2N3053†	60	40	5	700	1000	50-150	150	0.25	1.4	150/15	100				1 :	TO5
2N3704†	50	30	5	800	300	90-330	50		0.6	100/5	100	12			1	TO92a
2N3705	50	30	5	800	300	45-165	50		0.8	100/5	100	12			1 1	TO92a
2N3706	40	20	5	800	300	30-660	50		1.8	100/5	100	12				TO92a
2N3794	40	20	5	300	250	100 min	100	0.5	0.4	10/1	100	10				U29
40361†	70	70	4	700	1000	70-350	50	0.25	1.4	150/15	100	. •				TO5
40412	250			1000	1000	40 min	30				10				1 1	TO5

Medium Current Audio, Small Output Types, PNP

Type	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	hre hre	i _c mA	l _{cbo} μ A	V _{ce} (sat) @	I _c /I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
BC126	35	30	5	600	300	30-120	150	0.05	0.5	150/15	200t	5				RO97
BC160-10	40	40	5	1000	750	63 min	100	0.1	- 1		100t	20				TO5
BC303	90	65	7	1000	1000	40 min	150	0.02	0.65	150/15	60		l			TO5
BFX29†	60	60	5	600	500	50-200	10	0.075	0.4	150/15	100	12	ļ		}	TO5
BFX87	50	50	4	600	600	40 min	10	0.05	0.4	150/15	100	12			1	TO5
BFX88†	40	40	4	600	600	40 min	10	0.05	0.4	150/15	100	12				TO5
2N1131	50	35	5	600	600	20-45	150	1 1	1.5	150/15	50	45				TO5
2N1132	60	40	5	600	600	30-90	150	1 1	1.5	150/15	60	45			ļ l	TO5
2N3702†	40	25	5	200	200	60-330	50	0.1	0.25	50/5	100	12	i .		1 .	TO92a
2N3703	50	30	5	200	200	30-150	50	0.1	0.25	50/5	100	12				TO92a
2N4036	90	65	7	1000	1000	40-140	150	0.1	0.65	150/15	60		1		i .	TO5
2N4291	40	30	6	500	250	100-300	100	0.5	0.4	100/10	150t	10				U29
40362†	70		4	700	1000	35-200	50	-	1.4	150/15	100t		1			TO5

High Current Audio, NPN

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c max	P _{tot} max	h _{fe}	I _c mA	I _{cbo} μΑ	V _{ce} (sat)	I _c /I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
BD121	60	35	6	5A	303	30-100	1·5A	50	0.65	1A/100	95t					TO3
BD123	90	60	6	5A	303	30100	1.5A	50	0.65	1A/100	85t	l		ļ	1	TO3
BD124	70	45	6	4A	133	30-150	500	2	0.25	500/50	60	ĺ	1	1	ì	MD17c
BD130†	100	60	7	15A	666	2070	4A	5000	1.1	4A/400	1 · 1t			ļ	i	TO3
BD131†	70	45	6	4A	167	30-75	500			,	60	1	l	1	ł	X58
BD135†	45	45	5	1.5A	100	40-100	150	0.1	0.6	500/50	50	1	1	Ì	ì	X58
BD141	140	120	7	8A	666	20-70	2A		1	2A/200		1	Ì	1	1	TO3
BDY20	100	60	5	15A	i	20-100	4A		1.1	4A/400	1	!	t	ŀ	1	TO3
MJ481†	60	60	5	4A	500	30-200	1A	1000	0.4	1A/100	4	l	1	1		TO3
MJE521	40	40	4	3A	320	40 min	1A		,			1	İ	1		X58
MJE3055	70	60	5	10A	718	20-70	₫ 4A	1000	1.1	4A/400	2	Ì	i		1	X58c
TIP31 A	60	60	5	3A	320	20-100	1A .	500		,	3		Í	1	1	X75b
2N3054	90	55	7	4A	142	25-250	500	1000	1	500/50	0.75		ĺ	1	1	TO66
2N3055†	100	60	7	15A	660	20-70	4A	5000	1.1	4A/400	0.4		İ	1	1	TO3
2N4915	80	80	5	5A	500	25-100	2.5A	100	1	2·5A/250	4		1	l	1	TO3
2N5192	80	80	5	4A	320	20-80	1.5A	100	0.6	1 ·5A/150	2			į .	İ	X58
40250	50	40	5	4A	166	20-100	1 · 5A	1000	1.5	1 · 5 A / 150	1 · 2t		1	1		TO66
40251	50	40	7	15A	666	15-60	18A	5000	1.5	8A/800	0.5t					TO3

High Current Audio, PNP

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	l _c max	P _{tot} max	hre *hfe	I _c mA @	l _{cbo} μA	V _{ce} (sat)	I _c /I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz @	Base
BD132†	45	45	4	2A	167	35-75	500				60				1	X58
BD136†	45	45	5	1.5 A	100	40-100	150	0.1	0.6	500/50	50		l	l		X58
MJ491†	60	60	5	4A	500	30-200	1A	1000	0.4	1A/100	4				ľ	TO3
MJE371	40	40	4	3A	320	40 min	1A		1				ł	i	}	X58
MJE2955	70	60	5	10A	718	20-70	4A	1000	1.1	4A/400	2	Ì	1		1	X58c
TIP32A	60	60	5	3A	320	20-100	1A	500	1	1	3	1	1	1	ł	X75b
2N4906	80	80	5	5A	500	25-100	2.5A	100	1	2·5A/250	4	l I	į .			TO3
2N5195	80	80	5	4A	320	20-80	1.5A	100	1-4	1 5A/150	2)	1		1	X58
	"		1						1			!				}

See notes on power dissipation.

Medium Current Switching, NPN

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	I _c mA max	P _{tot} max	thre	I _c mA	i _{cbo} μΑ	Vc. (sat)	I _c /I _b mA	f _T MHz	C _{ob} pF	NF dB max	PG dB	MHz	Base
BSX20†	40	15	4.5	500	360	40-120	10	0.4	0.6	100/10	600			i		TO18
C762†	40	25	5	150	400	30-110	50	0.2	0.6	1 5 0/15	350	5		! }		TO18
MPS6531	60	40	5	600	310	90-270	100	0.05	0.3	100/10	390t	3.5	ì			TO92b
P346A	25	12	3		300	20 min	10	0.05	1	100/1	400	4		1		TO18
2N706	25	20	3	200	300	20 min	10	0.05	0.6	10/1	200	6		!!		TO18
2N706A	25	15	5	1	300	20-40	10	10			200	3.5	1	i i		TO18
2N2218	60	30	5	800	800	40-120	150	0.01	1.6	500/50	250	8				TO5
2N2218A	75	40	6	800	800	40-120	150	0.01	1	500/50	250	8	1	i l		TO5
2N2219	60	30	5	800	800	100-300	150	0.01	1.6	500/50	250	8		! !		TO5
2N2219A	75	40	6	800	800	100-300	150	0.01	1	500/50	250	8	Ì	1		TØ5

Medium Current Switching, PNP

Туре	V _{cbo} max	V _{ce} max	V _{eb} max	l _c mA max	P _{tot} max	hfE *hfe	I _c mA	I _{cbo} μA	V _{ce} (sat)	I _c /i _b mA	f _T MHz	C _{ob} pF	NF dB max	PG MHz dB	Base
MPS6534 V763† 2N2904 2N2904A 2N2905 2N2905A	40 25 60 60 60 60	40 25 40 60 40 60	4 5 5 5 5 5	600 600 600 600	310 400 600 600 600 600	90-270 30 min 40-120 40-120 100-300 100-300	100 50 150 150 150 150	0·15 0·05 0·02 0·01 0·02 0·01	0·3 1 0·4 0·4 0·4 0·4	100/10 150/15 150/15 150/15 150/15 150/15	250 300 200 200 200 200	6 8 8 8			TO92b TO18 TO5 TO5 TO5 TO5

_						
MD17c	MM10a.b	R09	R097	RO97a	TO1	тоз
(b;0)	1 2 3 a normal e b c b b e c	O dot	الله الله الله الله الله الله الله الله	□≡ □⊙∵	e o c Case isolated	Collector
T05	T07	TO18	T066	T072	T092	T092z
e c	eb c	b e Oc Case to collector	Collector	3 Case to s 1 2 3 4 a e b c 5 b b e c 5	1 2 3 a e c b b e b c	
T098	U29	X9a	x58	X58a,c	x59	X75
ecb	bce		ecb ecb metal pad to collector	ecb ecb metal/pad to collector	Cbe	bce mounting lug to collector



T is an unwritten law of electronics that whenever a group of engineers are gathered together then sooner or later a HiFi amplifier will be designed. Usually they, the amplifiers, that is, are assembled in such a way that even if they work well at the time they are quite impossible to reproduce.

The Texan is mainly a 'value-for-money' design which the average home constructor, armed with little more than a soldering iron, should be able to build successfully in a time which will allow him to remain on speaking terms with his wife.

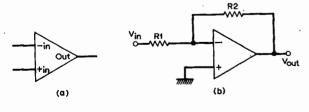
The original design was produced for a house-publication* and was intended to promote the use of integrated circuit operational amplifiers for audio use. Also, since I have an aversion to printed-circuit boards bristling with capacitors—particularly electrolytics—it was designed to reduce these to an absolute minimum.

The present circuit is updated to use the latest devices and it includes extra facilities such as scratch and rumble filters.

The printed circuit board has also been completely changed so that most of the controls are mounted directly and very little spagnetti wiring is required.

Why use operational amplifiers?

By definition the main function of a preamplifier is to provide gain and to do it in a way which is well controlled with respect to frequency response and stability.



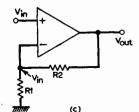


Fig. 1: (a) Triangle represents a device which has Infinite voltage gain.

(b) With two resistors added.

(c) Modified circuit—see text for detailed explanation.

If someone made a transistor with a potential voltage gain 200,000, an $f_{\rm T}$ of 2MHz, a maximum phase shift of 90°, differential inputs with high common mode rejection and so and so forth there would be little point in using an operational amplifier, but to the best of my knowledge they do not. However, operational amplifiers are now so easy to use that they can be treated simply as a transistor with eight legs.

More explicity the operational amplifier has the following features:

(1) A very high mid-band gain which allows accurate equalization while retaining sufficient loop gain to reduce distortion to very low levels.

(2) Inherently high input impedance and low output impedance which make feedback networks simple to design and minimise the effect of loading.

(3) Large potential output swing giving good overload ability.

(4) Balanced design with d.c. coupling and very low offset voltages. This greatly reduces the number of electrolytic capacitors required in the amplifier and allows the output to operate without the need for manual setting up of the d.c. output conditions.

(5) High supply ripple voltage rejection which means that the operational amplifiers will work on poorly regulated supplies and reduces the likelihood of low frequency instability, one of the most common faults in many amplifier designs.

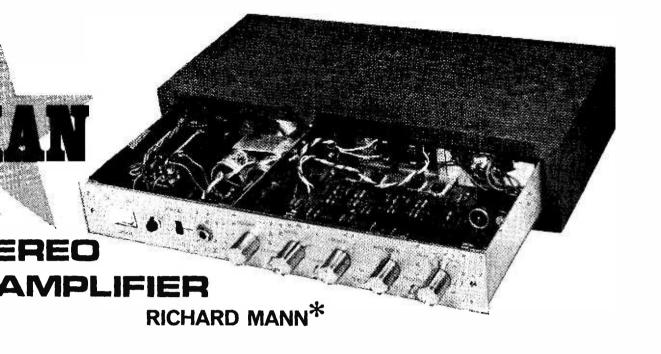
(6) Good common-mode rejection ratio of about 90dB. This describes the ability of the operational amplifier to see only the differential input and ignore voltages which are common to both input terminals so that spurious voltages appearing along a length of printed track, for example, can be almost totally rejected provided both inputs are referred to the same point on the track.

Operational amplifier theory

If you are familiar with operational amplifier theory skip the following. If not, read on.

Operational amplifiers are a very close approach, for most purposes, to the mythical 'black box' for which we make the following assumptions.

The triangle shown in Fig. 1a is a device which has infinite voltage gain (A) for differential voltages appearing across the $+_{\rm in}$ and $-_{\rm in}$ terminals, for which no current flows into the inputs and whose output impedance is zero.



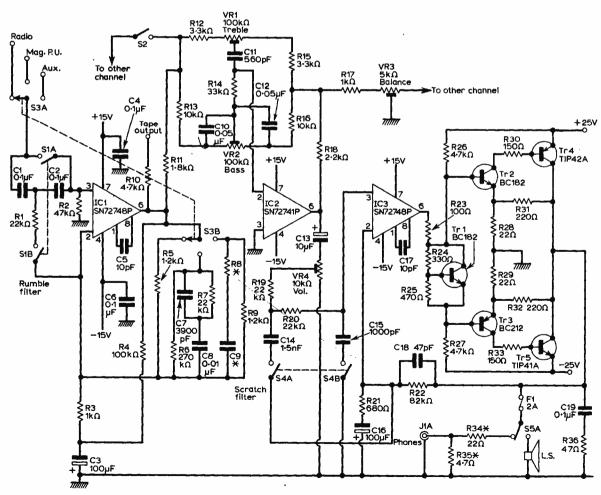


Fig. 2: One complete channel of the "Texan" stereo amplifier.

Therefore, if we add two resistors as in Fig. 1b, since the + input is earthed and $A = \infty$ there is no difference in voltage between the terminals and the - input stays at ground potential becoming a 'virtual earth'. The current through R1, must therefore be Vin/R1.

As no current flows into the 'box' all input current must flow through R2 producing a voltage across it of $\frac{V_{in}}{R_1}$ R2 and since the current flows from the virtual earth point, Vout must be

$$\left(0-\frac{V_{in}}{R \, l} \, R2\right)$$

Therefore, the closed loop gain of the circuit is

$$\frac{V_{out}}{V_{in}} = -\frac{R2}{R1}$$

If we modify the circuit as shown in Fig. 1c, we have a slightly different situation. The voltage at both input terminals becomes +Vin so the feedback

current is still $\frac{V_{in}}{R\, l}$ but it now flows in the opposite

direction so that $V_{out} = \frac{V_{in}}{R!}(R! + R2)$ and the closed

$$loop gain = \frac{V_{out}}{V_{in}} \left(\frac{R1 + R2}{R1} \right)$$

These two configurations are the classic 'inverting' and 'non inverting' amplifiers. The main differences, apart from polarity, are the input impedances, which for the former is R1 and for the latter is ∞ and the fact that the non-inverting amplifier cannot have a gain of less than unity.

The SN72748P

Most integrated circuit manufacturers make at least one version of the popular '741' operational amplifier design. This ingenious device was designed to improve on the equally popular '709' which in turn was a vast improvement on earlier integrated circuits which were essentially monolithic versions of discrete component designs. Although the 741 now has competitors from the new range of 'superbeta' amplifiers it is likely to remain an industry standard for a long time mainly because it offers very good all round performance at a very low price.

The SN72748P and SN72741P are fairly typical scions of the 741 family. The difference between them is merely that the SN72741P has internal frequency compensation which allows 100% negative feedback to be applied without the circuit becoming The SN72748P requires one small capacitor (about 10pF) to be added externally and gives a higher gain-bandwidth than the 741. Physically, they are identical in 8 pin dual-in-line plastic packages.

Typical parameters for the SN72748P when

operated at $\pm 15V$ supplies are:

Input resistance: $2M\Omega$ (higher with feedback. Output resistance: 75Ω (much lower with feedback). Gain: 200,000.

Input offset voltage (V_{IO}): lmV (this represents the matching error of the input transistors).

Maximum output current: 15mA r.m.s. Maximum output voltage: 9V r.m.s.

Input bias current (I_{IB}): $0.080\mu A$ (this is the current

* Specification

Sensitivity to give 20 watts into 8Ω Radio 30mV Magnetic pickup 2.5mV Auxiliary (see text)

Input impedance 47kΩ 1kHz

Tape output (low level output) 130mV via 4·7kΩ unaffected by tone or volume controls

Tone controls

Trebie +10 -12dB at 15kHz Bass ±16dB at 30Hz

Balance ±8dB one channel relative to other

Rumble—critically damped 2nd order corner frequency 50Hz, 12dB per octave roll-off

Scratch—critically damped 2nd order corner frequency 5.5kHz, 12dB per octave roll-off

Interchannel crosstalk

-51dB at 1kHz

-48dB at 10kHz

Unweighted signal-to-noise of complete amplifier with full amplifier bandwidth

-60dB magnetic pickup

-72dB radio

Dynamic range (equalisation and tone control stages) -38dB before clipping (above nominal input level)

Power Output (both channels)

20 \pm 20 watts into 8 Ω intermittent sine wave

16 +16 watts into 8Ω continuous sine wave

15 \pm 15 watts into 15 Ω continuous sine wave

Harmonic distortion

15 watts into 15 Ω 0.05% at 1kHz 20 watts into 8 Ω 0.09% at 1kHz

Low level distortion

0.16% at 1kHz 50mW into 15 Ω 0.07% at 1kHz 50mW into 8 Ω

Intermodulation distortion—wave analyser plots will appear later in this series

Frequency response (16 + 16 watts into 8 Ω)

-1dB 7Hz to 22kHz

-3dB less than 5Hz to 35kHz

Stability-will drive electrostatic loudspeakers

Output impedance less than 1 milli ohm

Dimensions (less teak sleeve)

width 363mm 14·3in height 48mm 1.9in

depth 149mm 5-9in

The following equipment was used to obtain the above measurements on the TEXAN amplifier:

Low Distortion Oscillator-Radford. Wave Analyser-Hewlett Packard 3590A/3593A. True R.M.S. Voltmeter-Hewlett Packard. Function Generator-Hewlett Packard 3300A/3305A. A.C. Digital Voltmeter-Pacific Measurements 1010. Oscilloscope-Hewlett Packard 181 A. Test Oscillators (two)—Hewlett Packard 652A (intermodulation tests). X Y Recorder-Hewlett Packard 136A.

necessary to turn on the input transistors). Supply ripple rejection: 20,000:1

More comprehensive details of the devices, including circuitry, can be obtained from manufacturers' data sheets.

Description of circuit

The complete amplifier circuit is given in Fig. 2, but components are referred to below for only one channel. In the components list, items for the other channel are +100 (i.e. R1, R101).

The input stage amplifies the various input signals to approximately 50mV, provides whatever equalization is required and gives the loop gain necessary for an active rumble filter. The feedback may look rather complex, but it breaks down into three parts as shown below.

(1) At d.c. the output is returned to the inverting input via R (Fig. 3a) which comprises R4 in parallel with R5, R6 or R9.

C3 blocks the path to ground so substituting in the equation for Fig. 1C we have a d.c. gain of $\frac{R+\infty}{\infty}$ (i.e. unity).

Since R2 refers the +ve input to ground, the -ve input will sit at a potential of $V_{\rm IO}$ (approx $\pm 1 mV$) and the output will also take this d.c. level. This offset is so low that there is no need for a coupling capacitor into the next stage.

(2) At mid band a.c. the gain of the input stage becomes $\frac{Z+R3}{R3}$ where Z (Fig. 3b) is the impedance

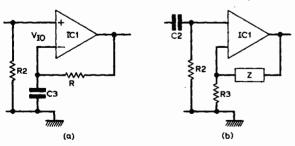
of the selected feedback network.

R4 has no effect on Z since the current through R4 is shunted to ground by C3. The feedback networks shown are designed to handle radio, magnetic pickup and auxiliary inputs and are discussed in more detail later in this series of articles.

The input impedance of the stage is inherently very high (>2M Ω) but is shunted by the $47k\Omega$ resistor R2 which provides the damping necessary for magnetic pickups and supplies the minute input bias current for the operational amplifier.

(3) A classic form of second order, high-pass active filter is shown in Fig. 3c.

The amplifier is ideally a unity gain buffer. Being a second order filter the low frequency response tends to roll off at 40dB/decade and the transitional region is a function of the damping factor (ζ) of the



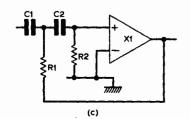


Fig. 3: (a) D.C. conditions of the input stage.

(b) At mid-band a.c. (c) Classic form of 2nd order high-pass active filter.

(see text for detailed explanation).

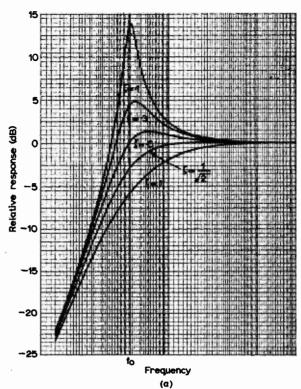


Fig. 4: (a) Theoretical response of 2nd order high pass filter for different values of ζ

filter. The effect of ζ is shown in Fig. 4a and has optimum cut-off without peaking when $\zeta = \frac{1}{\sqrt{2}}$. The damping factor is evaluated from $\zeta = \sqrt{\frac{R1}{R2}}$ and cut-off frequency (f₀) from f₀ = $\frac{1}{2\pi \sqrt{R1 R2 C1 C2}}$

Since the operational amplifier has to supply a certain amount of gain in the equalization stage its output obviously does not provide a unity gain buffer. However, the inverting input terminal can be used for this purpose since its voltage exactly follows that of the non-inverting terminal.

The characteristic of the filter is modified slightly because any current flowing down Rl, is not shunted to ground via R3 and C3, as one might think, but has to flow through the feedback impedance Z. Nevertheless the response is very close to a standard filter response as shown in Fig. 4b.

Placing the filter right at the front of the amplifier ensures that large inputs at sub-audio frequencies will not cause any intermodulation in IC1.

The push button switch S1 removes the filter when the button is out and a direct connection is made to the selected input.

This is permissable since the d.c. bias current to the operational amplifier is in the order of nano amps and causes no problems in the average pickup or tape head.

Facilities for coupling the equalised input signal to an external tape recorder are provided by taking the output of IC1 via a $4.7 \mathrm{k}\Omega$ resistor R10. This

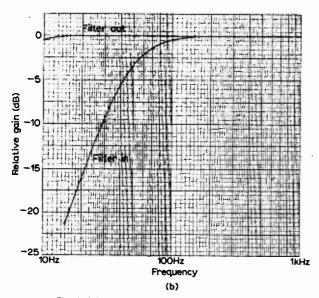


Fig. 4: (b) Rumble filter characteristic (measured).

series resistance allows the channels to be shorted externally if only a mono signal is required. The very low output impedance of IC1 (<1 Ω) means that no significant crosstalk is introduced in these circumstances.

The tone control stage

The tone control is a standard Baxandall circuit using an SN72741P to provide the loop gain. A fully compensated amplifier is necessary here since there is 100% feedback around the circuit in the cut position.

The time constants of the feedback network are chosen so that there is negligible interaction between the bass and treble controls. The control range is shown in Fig. 5.

The maximum levels of bass boost and cut are determined by the end stop resistors R13

and R16 respectively. Similarly the range of the treble control is limited by R12 and R15 although in this case the bass network does have a modifying influence which adds an extra limiting factor.

The transient response of the tone control with maximum treble boost is shown in the oscillogram, Fig. 6. It is well damped and shows no tendency to ring.

Two other functions are included in the tone control stage, the mono-stereo switch and the balance control.

Fnding the optimum position for the mono switch was rather a problem since it should be before

the balance control if possible. The series resistor R11 was added so that the two channels could be shorted together by S2 without overloading the preamplifiers (IC1 IC101). However, the resistor is not high enough to have a significant effect on the tone control range with the value given.

The balance control is an active circuit giving a control range of ±12dB for one channel relative to the other. This is a personal preference as I do not see much point in having 100% control range and it usually involves a waste of gain. However, anyone feeling strongly about it could cut out the relevant components and insert a dual gang pot between C13 and the volume control. The 10µF capacitor C13, incidentally, is the only d.c. block in the forward signal path which accounts for the good low frequency response and phase shift of the amplifier. A tantalum 'bead' capacitor is used in this position partly for its small size and partly because they will withstand a reverse polarizing voltage of 0.5V. Because the output voltage of IC2 depends mainly on its input offset it could be of either polarity but its magnitude will not exceed 200mV for any setting of the bass control (which determines the d.c. gain of the stage). A minor point here is that if the bass control is moved rapidly, a transient level shift may be heard from the loudspeaker. This effect can be removed by inserting a blocking capacitor of about 1µF between R13 and the junction of C10 and VR2, thereby fixing the d.c. gain of the stage at unity.

The scratch filter

The scratch filter is again based on a classic second order configuration as shown in Fig. 7. This low-pass circuit is the dual of the high-pass circuit used for the rumble filter (Fig. 3c). The

damping factor is $\sqrt{\frac{Cl}{C2}}$ and the cut-off frequency

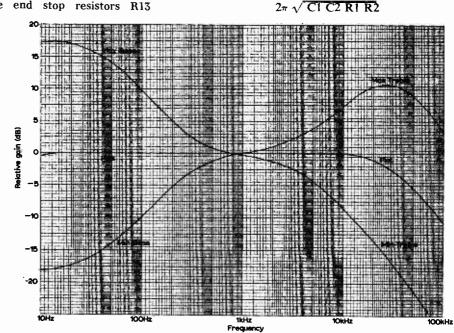


Fig. 5: The control range of the bass and treble circuit.

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20p 20p 20p 20p 20p

20p 20p 12p 10p 18p 20p

10p 10p 12p 12p

BC182L 10p BC188L 10p BC184L 12p BC212L 12p BC212L 12p BC213L 15p BC213L 14p BC213L 30p BCY32 30p BCY33 30p BCY33 45p BCY39 45p BCY39 45p BCY40 50p BCY40 50p BCY40 50p BCY70 15p

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BF18		OC71 18		7442 BCD-Decimal decoder (4-10-line) TTL O/P 759 789	70p 60p 55p	25 + 20p
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BFX		OC84 28 OC139 28	p 2N2369 10p	7470 Single J-K flip-flop (gated inputs) . 30p 27p	25p 22p 20p	500 + 9p
p BFX	86 25p	OC140 40	lo 15p	7472 Single J-K filp flop (gated inputs)	250 220 200	1000 + 8p
p BFX		OC141 66	D 2N2646 40p	7473 Dual J-K flip flop	85p 88p 30p 85p 88p 80p	
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BSX2	0 15p	ORP12 50	p 2N2906A		80n 75n 71n	N2926 Green 10p
D BSX2		ORP60 40 8T140 15		7490 BCD decade counter 75n 70n	38p 35p 33p 26 65p 60p 35p	25 + 9p
B8T9.	5A 12p	8T141 20	p 2N2907A	7491 8-bit shift register	90p 80p 70p	100 + 8p
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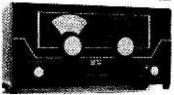
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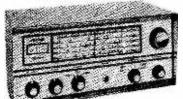
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Resistors:	Switches:
R1 R101 22kΩ 5% ½W R22 R122 82kΩ 5% ½W	
N N N 101 22822 3 /0 244	S1 4 pole 2 way S2 2 pole 2 way push-on—push-off
R1 R101 22kΩ 5% ‡W R22 R122 82kΩ 5% ‡W R2 R102 47kΩ 5% ‡W R23 R123 100Ω R3 R103 1kΩ 5% ‡W R24 R124 330Ω 5% ‡W	S3 4 pole 3 way make-before-break rotary
K3 K103 1K77 2 1 4 AA K74 K124 99077 2 4 4 AA	
R4 R104 100kΩ R25 R125 470Ω 5% ¥W	
R5 R105 1·2kΩ 5% 1W R26 R126 4·7kΩ R6 R106 270kΩ 5% 1W R27 R127 4·7kΩ R7 R107 22kΩ 5% 1W R28 R128 22Ω 5% 1W	S5 2 pole 2 way slide
R6 R106 270kΩ 5% ‡W R27 R127 4·7kΩ	S6 single pole mains on-off rocker
R7 R107 22kΩ 5% ¼W R28 R128 22Ω 5% ¼W	S1, S2 and S4 are for p.c. termination
R8 R108 — 5% \(\frac{1}{4}\text{W}\) see text next month	
R8 R108 — 5% ±W R9 R109 1·2kΩ 5% ±W see text next month	Integrated Circuits & Semiconductors:
R10 R110 $4.7k\Omega$ R29 R129 22Ω 5% $\frac{1}{4}W$	IC1 IC101 SN72748P)
R11 R111 1·8k $Ω$ R30 R130 150 $Ω$	IC2 IC102 SN72741P
R12 R112 3·3kΩ R31 R131 220Ω 5% ‡W R13 R131 10kΩ R32 R132 220Ω 5% ‡W	IC3 IC103 SN72748P
R13 R113 10kΩ R32 R132 220Ω 5% ‡W	Tr1 Tr101 BC182
R14 R114 33kΩ R33 R133 150Ω	Tr2 Tr102 BC182
R15 R115 3·3kΩ R34 R134 22Ω 10% 5W	Tr3 Tr103 BC212
R16 R116 $10k\Omega$ wire wound.	
R17 R117 1kΩ R35 R135 $4 \cdot 7\Omega$ 10% $\frac{1}{2}$ W	
R18 R118 2·2kΩ R36 R136 47Ω	Tr5 Tr105 TIP41A
R18 R118 2·2kΩ R36 R136 47Ω R19 R119 22kΩ 5% $\frac{1}{4}$ W R37 680Ω 10% $\frac{1}{4}$ W	D1 1N4002
	D2 1N4002
R20 R120 22kΩ 5% ½W R38 680Ω 10% ½W	D3 1N4002
R21 R121 08012 5% 4VV	D4 1N4002
All resistors 10% 1W high stability carbon film	ZD1 BZY88C15
unless otherwise specified.	ZD2 BZY88C15
Potentiometers:	
	Indicator:
VR1 VR101 100kΩ twin-gang carbon linear	N1 Neon mains indicator
VR2 VR102 100kΩ twin-gang carbon linear	
VR3 5kΩ single-gang carbon linear	Transformer:
VR4 VR104 10kΩ twin-gang carbon log	T1 Mains transformer primary 240V secondary
All pots have p.c. terminations—AB Metals type	20-0-20V 1A. (special design) Gardners SL20
D45	
Capacitors	Sockets:
C1 C101 0·1µF	SK1 5W DIN socket
C2 C102 0·1μF	SK2 5W DIN socket
C3 C103 100μF 3V tantalum	SK3 5W DIN socket
C4 C104 0·1μF	SK4 speaker DIN socket
C5 C105 10pF 10% 30V polystyrene	SK5 speaker DIN socket
C6 C106 0·1μF	J1 3 pole stereo jack socket
C7 C107 3900pF	
C8 C108 0·01μF	Fuses:
C9 C109 not normally required (see text, next	F1 F101 2A 1·25in. cartridge
month)	F2 F102 2A 1·25in. cartridge
C10 C110 0·05μF	F3 1A 20mm anti-surge
C11 C111 560pF 5% 30V polystyrene or	
ceramic	IC Sockets:
C12 C112 0.05µF	8 pin DIL IC sockets (IC sockets must be used-
C13 C113 10µF 16V tantalum	it is a condition of IC guarantee from Henry's Radio
C14 C114 1500pF 5% 30V polystyrene	Ltd.)
C14 C114 1500pF 5% 30V polystyrene C15 C115 1000pF 5% 30V polystyrene	
C16 C116 100µF 3V tantalum	ner 11
C17 C117 10pF 10% 30V polystyrene	Miscellaneous:
C17 C117	2 off twin fuseholders (1-25in fuses); 1 off panel
C19 C119 0·1μF	mounting 20mm fuseholder.
C20 1000μF 25V electrolytic Daly	All components, including a ready drilled fibre
C21 1000µF 25V electrolytic Daly	glass printed circuit board, drilled and punched
C22 3500μF 50V electrolytic Daly	metalwork, finished front panel, will be available to
C23 3500µF 50V electrolytic Daly	home constructors and the trade, from Henry's
All capacitors 10% 30V polyester or mylar, unless	Radio Ltd. A teak sleeve will also be available—
otherwise specified	more details next month.
a man triang a practical	

As with the rumble filter, its response is modified slightly by the feedback impedance of the output stage which provides its loop gain. The response of the scratch filter is given in Fig. 8 and shows that the circuit is again critically damped and reaches a roll off of nearly 40dB/decade(12dB/octave).

The output stage

The configuration of the output stage is almost identical to that of the input stage with a few extra watts thrown in. The midband gain of the stage is

 $\frac{R22+R21}{R21}$ and the reactance of C16 reduces this

to unity at d.c. so that the final offset of the power stage is only 3 or 4mV. This feature is of importance first because it removes the need for a level setting adjustment and secondly because it allows all normal loudspeakers to be connected directly without the need for a bulky blocking capacitor. Even a Quad electrostatic loudspeaker can be driven satisfactorily in spite of the low primary resistance of its matching transformer.

The power stage is rather unusual in that it provides about 20dB voltage gain unlike the normal Darlington or other 100% feedback configuration.

The gain is introduced by attenuating the feedback with R31 and R28 for positive excursions and

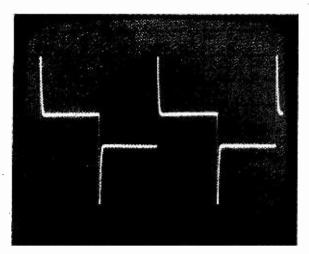


Fig. 6: Transient response of the tone control with maximum treble boost, Scale: X 200µS/div., Y 5V/div.

R32 and R29 for negative excursions. Voltage gain is necessary because the output swing of the operational amplifier is limited partly by its d.c. swing capability but mainly by its slew rate—the maximum rate in $V/\mu S$ at which the output can change.

The transistors used in the power stage are low cost plastic encapsulated types. The drivers are a complementary pair of Texas Silect devices type BC182 and BC212. They have a continuous rating of 300mW in free air at 25°C which is quite adequate for driving the output transistors providing ballast resistors R30 and R33 are added to reduce their collector dissipations.

Quiescent current setting

This is always a knotty problem in class B or class AB amplifiers and is a process which frequently means instant death to the output transistors. This does not matter too much if you work for the firm which makes them but it is not so good if you have parted with hard earned cash. For this reason the circuit shows fixed resistors (R24 and R25) which set-up the potential across the Vbe multiplier (Tr1). This potential, in conjunction with the d.c. feedback introduced by the emitter resistors R28 and R29, sets the current in the driver stage and hence, in the output transistors. If you are a purist you can sit down with a slide rule and data sheets and prove that this will give you a very wide variation of quiescent current. However, a more empirical

approach based on experience with about two dozen amplifier boards and transistors from different batches have indicated that the current usually sets up between 10mA and 40mA and anywhere in this range gives good crossover performance. This is partly due to the current, rather than voltage drive into the bases of the output transistors which have no shunt resistors to their emitters. Omitting these resistors slows down the power devices slightly but this is allowed for in the frequency compensation. The oscillograms in Figs 9a and 9b show the smooth crossover for an output power of 100mW r.m.s. with triangular inputs at 1kHz and 10kHz. Should the purist remain unconvinced, however, there is provision for a 470 Ω preset potentiometer on the printed circuit board. This is wired in place of R28 which should be omitted if the preset is used. There is also a link between the emitter of each TIP42A and the +25V rail which can be broken to give a current monitor point. This quiescent current in the output stage can then be set to 20mA.

Frequency compensation

It is vital to have an amplifier correctly compensated if good transient response is expected from an amplifier. All too often designs appear to have marginal stability which causes the home constructor endless problems. In the Texan a dominant pole is placed in the open loop response by the 10pF capacitor, C17. A further pole occurs naturally in the discrete stage which is cancelled by a zero introduced with the phase advance capacitor, C18, This gives nearly 90° of phase margin to the amplifier which allows it to cope quite well with highly capacitive loads. Fig. 10 shows the transient response when driving 20 watts into an 8Ω resistor in parallel with $2\mu F$ —said to be equivalent to an electrostatic loudspeaker load. There is a small overshoot under these conditions since transistor parameters have an inconvenient habit of changing with collector current. However, the degree of overshoot is not enough to cause trouble.

A similar oscillogram, Fig. 11, shows the output waveform when driving a Quad electrostatic loud-speaker, The power level in this case was only 1W as higher levels proved rather unpopular with my colleagues!

In addition to the compensation components there is a dummy h.f. load (C19 and R36) which is not necessary when driving most loudspeakers but takes

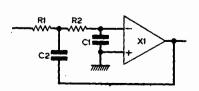
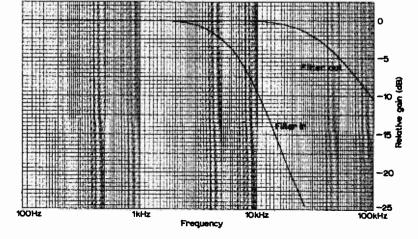
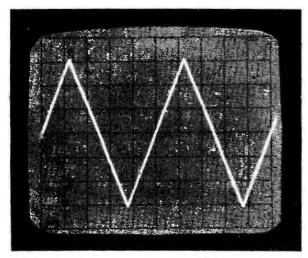


Fig. 7: (above) The scratch filter—classic 2nd. order configuration. Fig. 8: (right) Scratch filter characteristic.

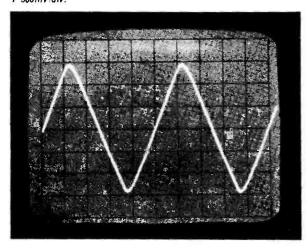




Oscillograms showing the smooth crossover for an output power of 100mW r.m.s. into 8 Ω .

9a (above): shows 1kHz triangular wave. Scales: X 200µS/div., Y 500mV/div.

9b (below): shows 10kHz triangular wave. Scales: X 20µS/div., Y 500mV/div.



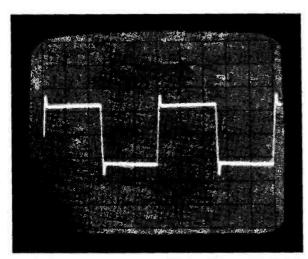


Fig. 10 : Transient response driving 20W into 8Ω in parallel with $2\mu F$. Scales : X 200 μS /div. Y 10V/div.

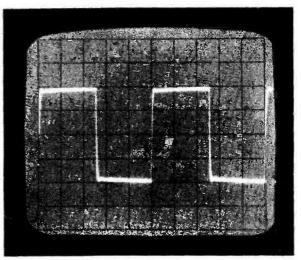


Fig. 11: Transient response when driving 1W into Quad electrostatic loudspeaker. Scales: X 200μS/div., Y 2V/div.

care of the output under nearly no-load conditions. These arise when using headphones or with some loudspeaker crossover systems which look very inductive at higher frequencies.

Headphone output

This facility was added at popular request and originally consisted of a jack socket straight across the output. An arrangement which promptly ruined a brand new pair of Koss PRO-4AA Stereo headphones not to mention a near miss on my eardrums. Although the former were kindly repaired free of charge, thereafter an attenuator seemed a good idea. The attenuator values were chosen to give a listening level very roughly equivalent to that through the loudspeakers and to be capable of driving normal low impedance headphones as those mentioned.

Overload protection

This is another brain-teaser. It is very easy to get so carried away with protection circuitry that it becomes as complex as all the rest of the amplifier put together. Simple methods frequently introduce distortion and do not protect much anyway. In the Texan the only protection is provided by a fuse. It gives satisfactory protection against a short circuit on the output line since the TIP41A and TIP42A have a continuous collector rating of 6 amps which allows an adequate margin for the 2 amp fuse (1 amp for 15Ω loads) to blow.

One disadvantage of using a d.c. coupled output is that a fault in the output stage could put a large continuous current through the speaker. The fuse will also protect against gross overloads but this is, of course, rather dependent on the actual loud-speaker used. It is very unlikely that a catastrophic breakdown will occur but if belt and braces are required then the traditional large electrolytic could be inserted providing the loudspeaker is returned to the -25V supply line to polarise the capacitor.

The power supply for a class B amplifier is the part most likely to be the subject of compromise. If size, weight and most of all, cost are of no consequence then a huge transformer and vast electrolytics will give almost perfect regulation. Alternatively, a regulated power supply can be built

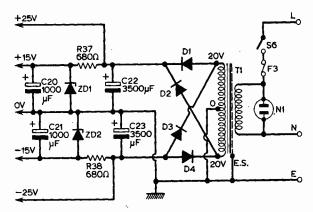


Fig. 12 Circuit of the mains power supply for the "Texan" stereo amplifier. Note that a 2A fuse (F2, 102) should be shown in the +25V and -25V supply rails respectively.

which cuts down on the electrolytics but adds at least one extra power transistor. Unless carefully designed they can also suffer from reactive output impedances or worse still instability which does not help the amplifier, of course.

The circuit diagram of the compromise chosen is shown in Fig. 12. For the output stage nominal $\pm 25V$ supplies are derived from an unregulated full wave circuit with 3500μF smoothing capacitors. The 25V rails in turn supply the ±15V rails for the operational amplifiers.

The 15V supplies are stabilised by a pair of low cost zener diodes. Accuracy is not at all important here since the operational amplifiers will operate equally well with supplies between $\pm 12V$ and $\pm 18V$. The zeners are necessary more as clamps to ensure that the rail voltages do not exceed $\pm 18 \bar{V}$, the maximum rating for the operational amplifiers. In parallel with the zeners are two 1000 pF capacitors, C20 and C21. These may seem rather high in value but with full bass boost and magnetic pickup equalization, an equivalent input in the order of 10 µV at 20Hz will produce 20W at the output. It is, therefore, essential to reduce ripple feedback to a minimum. The same reason, of course, accounts for the ease with which careless earthing can introduce low frequency instability.

The mains transformer recommended for the Texan is of rather unusual construction. It is a new design for which the manufacturers claim very low stray magnetic field-a great help when positioning the amplifier near a magnetic pick-up.

It also has a height of only 1.5in. which gives the Texan its 'slim-line' dimensions.

The nominal secondary voltage is 20-0-20 at 1 amp continuous r.m.s. and measurements on the transformer in circuit showed a regulation of about 3V r.m.s. between no load and a continuous sine power of 20 watts into one 8Ω load. The d.c. rails variations are as follows:

 $\pm 32V$ **Ouiescent** 20W sine into 8Ω (one channel only) $\pm 23V$ 16W sine into 8Ω

 $\pm 17V$ (both channels simultaneously) It is apparent from these figures that the power supply is a major influence on maximum output power and distortion when considering continuous sine wave inputs.

Although holders for the integrated circuits were not used on the prototype, it is however recommended that they are used by the constructor.

TO BE CONTINUED

ISSUES WANTED
...A May 1971 and two June 1971.—Martin Cooper, 1 Cranston Road, Bramcote Hillis, Beeston, Nottingham, NG9 3GU.
...P.W. May Issue 1967.—P. G. Cooper, 14 Ruspidge Road, Cindertord, Glos.
...P.W. August 1968.—P. Mitchell, 35 Inverton Road, Kirkcaldy, Fife, Scotland.
...Issue of P.W. which dealt with the construction of a continuous loop echo chamber (principle of record and fast/replay/erase tape recorder).—J. Peters, 83 Carlyon Road, Hayes, Middlessex.
...March 1966 issue of Practical Wireless.—M. Allmark, 4 Bowood Grove, Meanwod, Leeds, LS7 2PX.
action 1970, Feb. 1971, of P.W.—D. Knott, 39 Wychbury Road, Quarry Br., Briefrey Hill, Staffse, DY5 2XX.
Br., Briefrey Hill, Staffse, DY5 2XX.
Br., W. for August, September, October, November, All 1971.—F. Young, Church Hill, Holme Sp. Moort, Vark, YA4 4EE.
...danuary 1970 Issue P.W.—M. Brown, 47 Stanmore Road, Nahoon, East London South Practice Wireless Jan. 1971.—R. P. Leite, Caixa Postal 8984, S. Paulo 01009.

...January 1970 Issue P.W.—M. Brown, T. Commun. South Africa. South Africa. ...Practical Wireless Jan. 1971.—B. P. Leite, Caixa Postal 8984, S. Paulo 01009, ...Practical Wireless Jan. 1971.—B. P. Leite, Caixa Postal 8984, S. Paulo 01009,

...July 1971 Practical Wireless.—K. M. Warlord, 12 Bloomtreto Drive, Shemous, Bedfordshire. ...Issue of P.W. containing the "Clubman" receiver on a car radio.—A. Rubin, The Barn, Castle Lane, Todmorden, Lancs. ...Circuit of a 50watt Bass amplifier.—R. U. H. Parratt, 35 Cabell Road, Park Bank,

"... Circuit of a 50watt Bass amplifier.—R. U. H. Parratt, 35 Cabell Road, Park Bank, Gulldford, Surrey.

"Issues June to November 1970 inclusive and June 1971 of P.W. wanted. Please state price.—G. G. Cooper, 135 Eltham Road, West Bridyford, Norts.

"February '11 P.W. Expressly. Part 2 of P.W. Stereo tape recorder.—R. J. Searle, 2 Pontings Close, Blunsdon, Wilts.

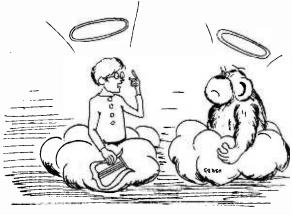
"A May 1984 Practical Wireless.—D. G. Robins, 21 Mountfield Road, Bramhalf. Stockport, Cheshire.

"Wanted for cast or loan, the two editions of P.W. at the end of 1970 or the beginning of 1971 that described the construction of a tape recorder.—Fit. Lt. Stewart, Cofficers Mess, RAF Brize Norton, Oxon.

"Wanted Issue of Practical Wireless daded February 1970, willing to pay full price.—M. J. Burrows, 7 Lyndhurst Road, Keynsham, Bristol.

"Wanted P.W. for: May 1967; Feb., March, Sept. and Oct. 1968; June 1969; Oct. 1970, and P.E. for Jay April 1967 inclusive line. supplement for April Issue preferably. Also, borrow P.W. Sept. 1967.—A. Blay, 28 Bourne Close, Broxbourne, Herts.

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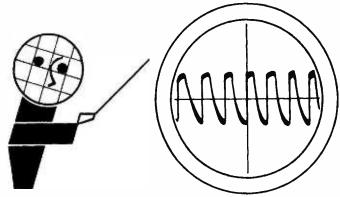


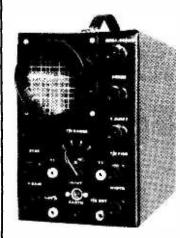
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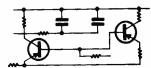
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OR2 OZ4	.30 6BZ6	.31	68A7M	. 85	128H7	.15	85 A 2	.48	EBC41	. 48	EL37	.74	MHLD6 .7		PL504/500	UF80	.85		.13	BCY34 BCY38	. 23	OA200 OA202	.09
1A7GT	.25 6CB6A		68G7GT			. 28	90C1 150B2	.58				.53	N78 2.0		69		. 34	AC127	.17	BCY39	. 25	OC22	.38
1B3GT	. 35 BC9	.78	68H7		128070		807	.59				.58		4	PL505 1.80 PL508 .90		.63	AC128 AC154	.20	BCZII	. 38	OC23	.38
1H5GT	.88 6CD6G			.35		. 50		-80				. 38	P61 .4	14	PL509 1.80	UL41	.54	ACI56	.20	BF158 BF159	.25	OC24 OC25	.38
INSGT	.18 6CG8A		68K7GT 68Q7GT			· 48	5763 AC2/PE	. 50	EBL21	. 60		. 22	PABC80 .3 PC86 .4		PL802 .75		.31	AC157	. 25	BF163	. 20	OC28	. 88 . 60
1R5	.28 6CL6	.43	6U4GT	. 60		.24	102/11/	. 98		. 59		-38	PC86 .4 PC88 .4		PM84 .34 PX4 1.18		. 33	AC165 AC166	.25	BF173 BF180	.38	OC29 OC35	.68
184 1U4	.22 6CM7		6V6G 6V6GT		19BG66				EC92	. 84		. 23	PC95 .5	8	PX 25 .98		. 25	AC168	. 38	BF181	.40	OC36	.43
1U5	.48 6CW 4		6X4	. 20	1966 20 D4	.50 1 · 05	ACSPEN	. 98		1.50		. 32	PC97 .3 PC900 .3		PY33/2 .50 PY80 .33		.45	AC176	. 55	BF185	.40	OC42	. 63
2D21	.35 6 DE7		6X5CT		20F2	. 65		(7)	ECC40	.60		.75	PCC84 .2		PY80 .33 PY81 .24		1.73	ACV17	. 28	BFY50 BFY51	.23	OC43 OC44	1 - 18
2GK 5 3A4	.50 6DT6A		7A7 7B6	. 88	20L1 20P1	.98	AC/TP	.98		-16		. 38	PCC85 .2		PY82 .24	U22	. 39	ACT18	.20	BFY52	.20	OC45	.10
3D6	.19 6F1		7B7		20P3		AL60	. 78		. 19		. 75	PCC88 .4 PCC89 .4		PY83 .26 PY88 .32		. 64	ACY19	.19	BFY194	- 15	OC46	.15
3Q4 3Q5GT	.88 6F6		7C6		20P4	. 88	ATP4	. 12	ECC84	28	EM84	.31	PCC189 .4		PY500 .95		.78	ACY21	.19	BY100 BY105	.18	OC70 OC71	.13
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3V4	.82 6F14	.42	7Y4	. 60	25 L6G		AZ41	.53		.35		.35	PCF82 .3 PCF84 .4		PY801 .33 PZ30 .48	U191 U251	. 58	ACY28 AD140	.18	BY126 BY127	.15	OC74	. 23
4CB6 5CG8	50 6F18 50 6F23	. 45			25 Y 5 G		CV 63	. 90			EY81	. 35	PCF86 .4	4	QQV03/10	U281	.40	AD149	.50	BYZIO	.25	OC75 OC76	. 11 . 15
5V4G	84 6F24	. 68	10C2		25Z4G		CYIC	. 58 . 53			EY83 EY84	.54	PCF200 .6' PCF801 .2		1.20 Q895/10 .49	U282 U301	. 40	ADIGI	. 45	BYZ11	. 25	OC77	.27
5 Y 3 G T 5 Z 4 G	.26 6F25		10DE7 10F1	. 50	25Z5 25Z6G		CY31	. 31	ECF80	.27	EY87/6	. 30	PCF802 .4		QV04/7 .63	U403	.83	AD162 ADT140	.63	BYZ12 BYZ13	. 25	OC78 OC78O	.15
6/30L2	. 55 6F32		10P9	45			DAF91 DAF96	. 20 . 33		. 26	EY88 EY91	.40	PCF806 .5: PCH200 .6:		R10 .75	U404	. 38	AF106	.50	CG12E	. 20	OC81	.11
6A8G 6AC7	.33 6GH8A .15 6GK5		10F18	. 35			DF91	.14	ECF804		EZ40		PCL82 . 35		RII .98 RI6 1.75	U801 U4020	.93	AF114 AF115	.25	FSY11A FSY41A		OC81 D OC82	.11
6AG5	.25 6GU7		10LD11 10P13	. 53	30C17 30C18		DF96 DH76	. 34 . 28	ECH21	2 · 10 · 68	EZ41 EZ80		PCL83 .58		R17 .88	VP23	.40	AF117	.19	GD9	. 20	OC82D	.11
6AK6	.30 6H6GT	. 15	10P14 1	.08	30F5	. 65	DK 40	. 55	ECH 42	.60	EZ81	.22	PCL84 .34 PCL805/85		R19 .30 SP42 .75	VP41 VT61A	.38	AF121 AF124		GET113 GET116		OC83	. 20
6AM8A 6AN8	.50 6J5G .49 6J5GT		12A6 12AC6		30FL1 30FL2		DK92 DK96	. 35	ECH81		FW4/500	.75	.40	0	SP61 .33	VUIII	44	AF125		GETIIS		OC84 OC123	.24
6AQ5	.22 6J6	.18	12AD6	.40	30FL12	. 69	DL96	. 85 . 85	ECH84		FW4/800 GZ30		PCL86 .38 PD500 1.44		TH4B .50 TH233 .98	VU120 VU120	.60	AF126	.18	GET119	.20	OC139	. 23
6AR5 6AT6	.18 6J7GT		12AE6 12AT6		30FL14		DM70	. 80	ECL80	. 80	GZ32	.41	PEN4DD	1	TP2620 .98	VU133	. 35	AF139 AF178	. 65	GET573 GET587	.38	OC140 OC169	.95 .23
6AU6	.20 6JUSA	-50	12AT7	.16	30L17	. 67	DM71 DY87/6	. 38 . 24	ECL83	. 52		.70	1.38 PEN45 .40		UABC80.80 UAF42 .49	W107	.50	AF180	.48	GET873	.15	OC172	.85
	.28 6K7G		12AU6 12AU7	.21	30P4MR	.95	DY802	.85	ECL84	.54	GZ37	.67	PEN45DD		UBC41 .45	W729 X41	. 50	AF186 AF239	.55		23	OC200 OC201	. 22
6AX4	.39 6K8G				30P12 30P19/	. 69	E80F 1 E83F 1		ECL85		HABC80 HL23DD		.76 PEN453DD	5	UBC81 .40	Transist	BTO	BA102	.45	GET898	. 23	OC202	.48
	.18 6L1	- 98	12AX7	. 22	30P4		E88CC	60	EF22	. 63	HL41DD	98	. 98		UBF89 .30	& Diode 2N404		BA115 BA116		M1 OA5	.15	OC203	.30
	.20 6L6GT		12BA6 12BE6					40 90	EF40 EF41		HL42DD HN309 1		PENDD-	11	UBL21 .55	2N2297	.23	BA130	.10	OA9	. 28	OC204 OC205	.30
6BE6	.21 6L18	.44	12BH7	.27	30PL14	. 65	E182CC1		EF42	. 33			4020 .88		CC92 .85	2N2369		BA153		OA10		ORP12	.53
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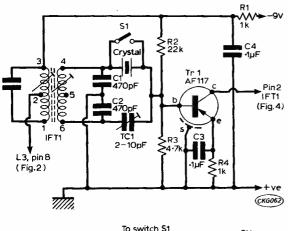
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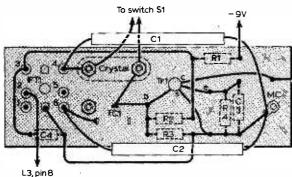
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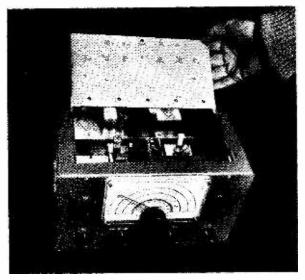
Fig. 6 is the circuit of this unit and if fitted is placed between the oscillator coil L3 (Fig. 2) and the i.f. amplifier described. The filter i.f. transformer provides a balanced output at pins 4 and 6, TC1 being adjusted to compensate for stray capacity in the crystal and parallel switch. In these circumstances selectivity can be extremely high but as the crystal works into a moderately high impedance here the passband is wide enough for a.m. reception without too much deterioration of speech quality.

When the switch is closed, the crystal is out of use, and the selectivity is that provided by the four i.f.t.'s.

The filter i.f.t. has to be modified, to obtain a centre-tap (pin 5 is unsuitable). Pull the screening can off the i.f.t. A small capacitor is wired internally between pins 4 and 6 and this component is snipped off. (The similar capacitor between pins 1 and 3 must be left). The i.f.t. is then replaced in its can. The parallel capacitance is now provided by the two capacitors C1 and C2, wired externally.







Construction. The filter is assembled complete on a piece of paxolin or Veroboard 3×1 in. as in Fig. 7.

A black flying lead (or pin) is provided for the negative connection. The unit is mounted by two lin. bolts, with extra nuts which provide the chassis return. When the unit is fixed to the receiver chassis, the lead from pin 8 on the oscillator coil is near the filter i.f.t., and can be soldered to pin 2. The lead from pin 2 of i.f.t.l, (Fig. 4), is now soldered to the collector lead of the filter unit transistor or to a pin used as anchorage.

A hole must be drilled in the chassis to reach the lower core of the i.f.t. Two insulated leads are soldered to the crystal holder sockets passing directly down through holes to the crystal switch.

The crystal switch is fixed directly under the crystal. It can be a rotary switch, fitted to a bracket, with a coupling and extension spindle. Alternatively a switch wafer can be attached to the back chassis runner with 12in. spacers and operated by a 14in. dia. polystyrene rod cut to fit.

Filter Adjustment. A signal generator and output meter are helpful, though adjustment is possible by ear. The highest selectivity obtainable is very sharp so having adjustments a little off maximum is an advantage. Adjustment may need some patience, depending on how close the crystal and i.f.t. frequencies match at the beginning.

If alignment is badly out, no signals at all may be heard. If so, align the i.f.t.'s as described earlier, switch the crystal out (switch closed) and adjust the two filter cores for best volume. Set the phasing capacitor TCl about half closed.

Fig. 6: (top left) shows the circuit for the optional crystal filter, IFT1 being modified as described in the text. Fig. 7: (left) gives the layout of the filter on its 3 x 1in. board.

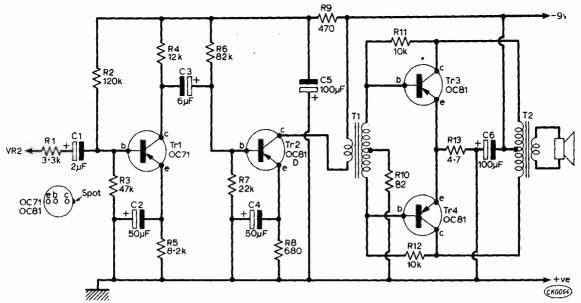


Fig. 8: Circuit of the audio stages, fed from the detector stage (Fig. 4, Part 1) via the volume control VR2 mounted on the front panel.

Open the crystal switch, and tune the receiver (or generator). If results are much as before, crystal and i.f.t.'s are by no means on the same frequency. Assuming no generator or other means of checking the frequency of the i.f.t.'s is available, switch the crystal out, adjust all seven i.f.t. cores about a half turn in the same direction, and re-peak them for best results. If bringing in the crystal now sharpens tuning, the shift made to the intermediate frequency was correct. If results seem little changed or worse, the shift was the wrong way, so all the cores have to be moved in the opposite direction.

When crystal and i.f.t.'s are approximately in agreement, tuning through a signal (or tuning the generator) will show two tuning points. One is extremely sharp, and is not changed by adjusting the cores of the i.f.t.'s in later stages. The other is flatter, and can be moved by rotating the cores. The sharp peak is the crystal resonance. Tune to it and then rotate each i.f.t. core to the same frequency, as indicated by a great increase in sensitivity and the eventual disappearance of the second, flatter response.

With i.f.t.'s aligned to the crystal frequency, tune through a signal (or tune the generator) while

adjusting the phasing capacitor TC1. If TC1 is completely unscrewed, a very sharp null will be found at one side of the signal, the signal re-appearing as this is passed. If TC1 is fully screwed down, the null arises at the other side of the signal, so it is adjusted to give a reasonably even response. Adjustments may make necessary very slight readjustment of the filter cores.

AUDIO AMPLIFIER

With an unlucky combination of values in the capacitors C1 and C2 the error could prevent phasing. If so, place a 30pF pre-set across C1 or C2, re-adjust the core, and adjust as described. It may be found that exchanging C1 for C2 will suffice. If selectivity worsens with the pre-set at any value, remove it and put it across the other 470pF capacitor.

The circuit for this is shown in Fig. 8. It requires a few more components than directly-coupled circuit but has the advantage that d.c. operating conditions for one stage do not depend on those of other stages, so that unsatisfactory working is not likely, while all items are easily obtained.

Construction. The parts are assembled on an insulated panel $3^{1}_{2} \times 2$ in. as in Fig. 9. The amplifier can be teasily tested by adding a speaker, 9V battery and providing some form of input, such as a pickup, or by connecting it to the receiver volume control.

The amplifier is mounted by two $^{1}2in$. 6BA bolts, with extra nuts, which also secure tags, and act as a return to the receiver chassis. In the event of the amplifier drawing a very small current with no signal (say under 3mA) and exhibiting cross-over distortion, the 82 Ω resisfor R10 may be increased slightly in value, 100Ω being suggested.

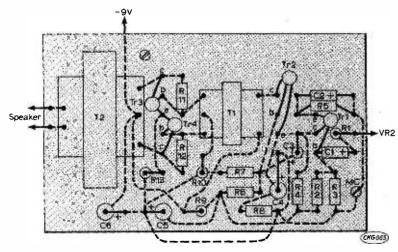


Fig. 9: Layout of the audio amplifier constructed on a 3½ x 2in. insulated board. It is mounted on top of the chassis with T2 at the rear, see Fig. 12 and photographs.

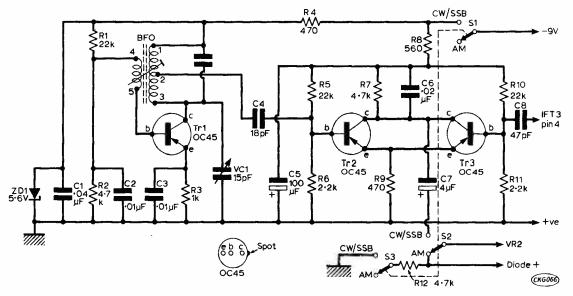


Fig. 10: Circuit of the b.f.o.|product detector required for the reception of c.w. and s.s.b. The function switch S1, 2, 3 is mounted on the front panel.

BFO/PRODUCT DETECTOR

The receiver as described so far is intended for a.m. reception and is thus suitable for all normal short wave listening. In some cases it may be wished to receive Morse and single sideband transmissions, especially on the amateur bands. This can be done by using a beat-frequency oscillator/carrier oscillator, with product detector.

The circuit for this unit is shown in Fig. 10. Tr1 is the oscillator, operating at 1.6MHz. This frequency is varied by the panel pitch control VC1. The zener diode ZD1 provides a stable voltage for the oscillator.

Output from the oscillator goes to one of the pair Tr2/Tr3, with c.w. or s.s.b. signals fed in from the i.f. amplifier via C8.

For c.w. reception, VCI is adjusted so that the required audio heterodyne note is produced.

For s.s.b. reception, Trl operates as carrier oscillator, to replace the carrier removed during s.s.b. transmission, thus allowing the s.s.b. signal to be resolved.

A rotary function switch provides for c.w./s.s.b. reception or a.m. Section S1 interrupts the negative line for a.m. reception, while section S2 switches the volume control VR2 to the product detector circuit, or existing a.m. diode detector, as required. Section S3 brings in R12 for c.w./s.s.b. to complete the diode circuit made through VR2 during a.m. reception.

Construction. The components are assembled on an insulated board as in Fig. 11. All components, including the b.f.o. coil, are on this side of the board, so it must be checked that coil connections are correct when turning the board over.

Flying leads are provided for external connections and the board mounted clear of the chassis by ¹2in. bolts with extra nuts. The point MC is a tag locked to the board thus completing the circuit to the metal chassis when the assembly is mounted in place.

Capacitor C8 is near the final i.f.t. a lead being soldered to pin 4 of the latter and passing through the chassis to C8.

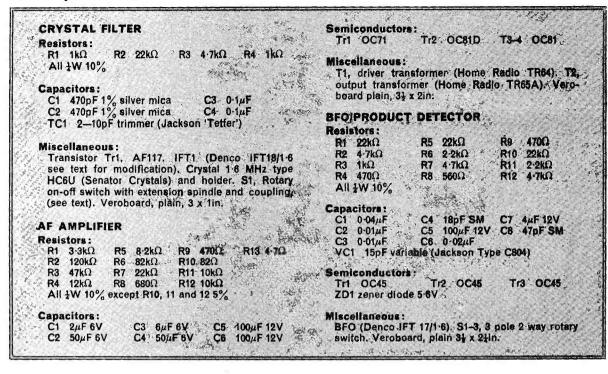
A black lead from R4 is connected to S1, which completes this circuit for c.w./s.s.b. reception only. VR2 is disconnected from the diode circuit and

taken to S2. This connects to C7 in the c.w./s.s.b. position and to the diode (as originally) for a.m. reception. R12 is wired directly to the switch as in Fig. 10.

Alignment. With the switch in the a.m. position, reception should be as before but the final i.f.t. core will require a slight adjustment for best results.

Set VC1 half closed, switch to c.w./s.s.b., and rotate the b.f.o. coil core until a strong heterodyne is heard. Leave the core at the central 'zero beat' position. Turning VC1 either way will now produce a heterodyne which rises in pitch.

Fig. 11: Layout of the b.f.o.|p.d. which is mounted under the chassis below the audio amplifier board.



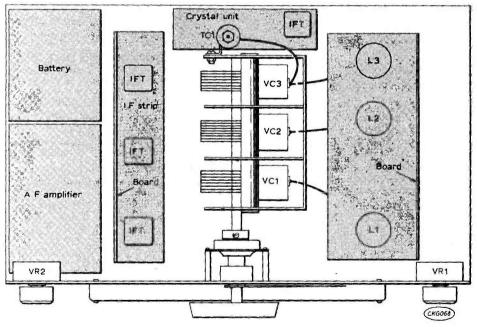


Fig. 12: Location of four of the units and the battery, the b.f.o.|p.d. being mounted underneath. Compare with photographs. An aluminium screen was later fitted between L1 and L2 on the r.f. unit, to improve stability. This can be seen in the photographs. Mounting instructions are included with the tuning dial.

When a c.w. signal is tuned in adjust VC1 for best reception the b.f.o. being above or below the signal, whichever results in least interference from other signals.

When receiving s.s.b., VC1 is carefully adjusted to give best resolution of speech. If resolution is impossible, the oscillator is being tuned to the wrong side band of the s.s.b. signal. The same sideband is nor-

mally in use on any one amateur band, so the adjustment of VC1 will hold good for that band.

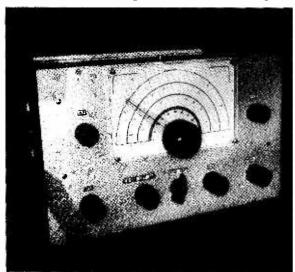
With s.s.b. it is normally necessary to reduce r.f. gain somewhat by operating VR1. If the crystal is in use, tune in the s.s.b. with the switch at a.m., then switch to s.s.b. and adjust VC1 for best resolution. VC1 can then be generally left untouched while tuning the band for other s.s.b. signals.

GENERAL CONSTRUCTION

Connections and placement of controls, etc., are as follows:

VR1. This is fitted at the right of the panel. Leads from the r.f. and i.f. units go to one outer tag. The centre and other outer tags are wired to chassis.

VR2. This fits at the left. ZD1 positive runs to one outer tag, the centre tag goes to the a.f. amplifier, and the other outer tag to chassis. (If rotating the



Panel layout. Top left, Volume control and top right, r.f. gain control. Left to right below:-b.f.o. tuning, function switch, crystal filter in/out, mixer and aerial tuning.

knob gives reversed control of volume, change over the leads to the outer tags. This also applies to VR1.) Connect one switch tag to chassis, and a red lead with positive clip to the other switch tag.

VC4. This is located between the crystal switch and VR1, and is connected to a lead from VC1. VC5 is similarly connected to VC2. Rotor tags are earthed to an adjacent tag bolted to the chassis.

Negatives. An insulated tag is bolted underneath the chassis near the a.f. amplifier, and is an anchor point for battery negative, and the negative leads running from each unit.

Aerial-Earth. Two sockets are fitted to the rear chassis runner. That for earth is connected to the chassis. The lead from pin 8, L1, runs to the Aerial

Speaker. The secondary leads from T2 run to separate sockets, or a jack outlet. A 2/3 ohm speaker is required, fitted in a cabinet or attached to a baffle board. For c.w. reception low impedance headphones should be used.

Panel. This is secured to the chassis by the control bush nuts and two brackets. The tuning drive is fitted by the method described in the instructions supplied with it.

Cabinet. Holes are punched in the back, level with the sockets. An opening lid is made in order to reach the coils and the battery, which rests on the chassis near the a.f. amplifier. Small holes are drilled close together in order to start a metal-saw, and a piece about 10×5in. cut out. Two strips of metal about 6×1in. are fixed inside with countersunk screws, for the lid to close against. It swings on a 10in, hinge bolted at the back.

CO! CO! CO! CO! CO!

...can anyone suggest a suitable speaker for use with my "Mellophone 3" set. It is unuted in a smart mahogany cabinet standing on legs. Valves used are PM2DX, PMLE (2), and three PM2A's.—H. Owen, 34 Bentgate Street, Newhey, Rochdale,

Lancashire.
...can anyone please lend me a copy of Modern Wireless No. 1.—L. A. Slack, 60 Parsonage Road, Withington, Manchester, M20 gWQ.
...I have recently acquired a Cariton SG 4 receiver about 40 years old and I cannot find any gen on it at all (valves, types, etc.). Please can anyone help.—A. C. Lintott, 3 Seaham Street, Silksworth, Sunderland, SR3 1EX.
...details and copies of radio circuits from early editions of radio magazines, as I wish to renovate some Vintage receivers.—S. Laffan, St. Kieran's College, Kilkenny Ireland

...any gen on the µA 703C i.c.—J. Rowley, 18 Old Park Road, Shirehampton, Bristol, BS11 UPW.

...any gen on the μA 703C i.c.—J. Rowley, 18 Old Park Road, Shirehampton, Bristol, BS11 LPW.

...any gen on the Rx type R1475 p.s.u. type R360 and tuner unit type 131.—C. Astbury, 16 Eaton Avenue, Handbridge, Chesier, CH4 JHB.

...any gen on R4187 S.T.C. set 17 valve, 24 channel receiver circuit. Manual and mods for mains. Loan or buy, Also back nos. P.W. for last 10 years. Sell at cover price plus postage.—K. Armstrong, 11 Woodville Gardens, Rulstip, Middx.

...circuit and details of R209 Rx. Also gen on B2 (spy) receiver and layout if possible.

I. L. Bishop, 27 Stratton Way, Biggleswade, Beds.

...any gen on Reiner Electronics Oscilloscope No. 556. Service Manual, modification details etc. Will return.—R. P. Davies, 61 Kenfey Road, Kingston, Surrey.

...circuit or Information on capacitor analyser (Solar) type CB-2U. Loan or purchase.

R. Kirk, 22 Barncraig Street, Buckhaven, Fife, Sociland.

...circuit diagram and gen on the R1155 Rx. I will copy and return.—A. McKenzle, 6 Moorfield, Newcästle upon Tyne, 2.

...service sheet for Pamphonic Reproducers Ltd. type 3000 stereo amplifier.—

R. W. Parker, 11 Greenacres, Downton, Sallsbury, Willts.

...any gen on "Contessa" receiver which is 5 to 8 years old with 6 transistors and a diode and what appears to be 2 Lif. stages. The tuning dial contains the word Contessa and at the bottom of the speaker fret it reads "transistor".—H. Gosling, 12 Ullswater Avenue, Halfway, Sheffield.

...any Info on making a rumble and scratch filter.—N. Davidson, Learnington House, Dover College, Dover, Kent.

...any hints and mode of the Atlantic, St. Donat's Castle, Llantwit Major, Giamorgan, CF6 9WF.

CF6 SWF.

...Instruction manual for Airmec 877 Televet test set.—D. Wigley, 4 Sollons Avenue, Harlesden, London, N.W. 10.

...any Info on converting the 19 set Tx/Rx.—M. Saxon, Shepherds Fennel, Hendon Wood Lane, London, NW7 4HR.

...circuit for a.m./f.m. tuner with a multiplex system.—A. G. Sobers, c/o Donnachie, 181 Octavia Court, Greenock, Renfrewshire, Scotland.

...circuit and any Info on constructing an echo (depth) sounder.—R. Mediand, 10 Northfield Way, Brighton, BNI 8EH.

...blueprint or practical circuit for the 4-transistor amplifier published in P.W. November 1981.—B. Endean, 25 Astley Gardens, Seaton Sluice, Whitley Bay, Northumberland.

...circuit diagram of an electronic echo chamber: the type which does not employ ... circuit olagam or an electronic echo chamber; ne type which does not elipioy an endless tape with multiple heads but one which will produce both sustained echo (reverb) and repetitive echo with a variable time factor.—M. Hale, 27 Rachel Gardens. Selly Oak, Birmingham, 825 6NY.

... the circuit diagram of a stylus organ or even the details of the one published in P.W. some years ago.—V. P. Kelly, L/103197 REM(A) Mess 6F1, H.M.S. Ark Royal, BFPO, London.

BFPO, London....blusprints for the following: "Celeste 7-transistor portable" (June 1963), "'Mc/s
Transcelver" (June 1964), "Beginners S.W. Superhet" (December 1964) 'Progressive
S.W. Superhet" (February 1968) and "Mini-Amp" (November 1961).—A. Raza,
P.O. Box 2671, Tripoli, Libya.

c.circuit or manual for TS210, to loan or buy. Also, any useful mods would be
welcome.—D. Millard, 20 Jubilee Road, Tipton, Staffs, DY4 0QP.

BOOKS WANTED

...copies of the Marconigraph, or any other old radio mags. pre 1930. Also a nice set of Harmsworth's Wireless Encyclopedia.—C. Riches, Practical Wireless Editorial Dept., Fleetway House, Farringdon Street, London, EC4A 4AD.

CORRESPONDENTS WANTED

...any one interested in electronics whether it be ham radio, Hi-Fi, audio, TV o any other aspect.—Des. Walsh, E15CD, "Coombe Down", Ballylynch, Carrick-on Sulr, Co. Tipperary, Ireland.
...P.E. May and July 1970. Will exchange for P.E. September to October 1989.—Des. Walsh, "Coombe Down", Ballylynch, Carrick-on-Suir, Co. Tipperary, Ireland

ISSUES WANTED

..1950-1955 and Radio Const. 1950-1959.-Lurton, 12 The Vale, Acton, London, Feb. and March 1971 P.W.—R. Watson, 51 Woodbridge Road, Newbourne,

n. Feb. and march 1971 - W. H. Watson. 31 Vocabriage Road, Newdourne, N. Woodbridge, Suffolk. N. Revision 1970 - Woodbridge Road, Newdourne, N. Television for Oct, 1987, Feb. March, April, May, June, July 1989 and Dec. 1970.—P. F. Johnson, 21 Juliliee Drive, Sheringham, Norfolk. March, Radio Amateur Sept. 1953, RSGB Sulletin February 1953, S. W. Mag. Oct. 1948 and Dec. 1984.—I. Simpson, 2 Taylor Cottages, Derryhale, Portadown, Co. Armagh.

N.I.

"Jan. to Nov. 1966, July 1967, Jan., Feb., March 1968, Index for Vols. 41 to 46 (all P.W.), P. Electronics for Jan. to Nov. 1966, Nov., Dec. 1967, Jan., March 1968, Jan. to Nov. 1969, Nov. 1970, April 1971 and index for Vols. 1 to 6.—H. Ferns, 11 King Street, Coatbridge, Lanarkshire, Scotland.

"Issue of P.W. Jan. 1982 containing "Minuette".—S. Grosvenor, 50 Countess Crescent, Bispham, Blackpool, F72 9 LQ.

"P.W. for June and July 1969 also Jan. 1971. P.E. issues Jan. to March 1970 inclusive.—A. Taylor, 89 Howth Drive, Woodley, Nr. Reading.

"P.W. September and October 1970.—W. R. Pitt, 26 Highdown Avenue, Worthing,

Sussex.

W. June 1971. F. J. Camm's "Beginners' Guide to Radio" (5th or 6th edition). Also "Elementary Course in 28 Lessons", and P.W. July 1948.—H. Howard Shepherd, 5 Pafford Avenue, Watcombe, Torquay, TQ2 8BS.

"The Issue of P. Television containing the "Viewmaster".—E. Farrer, 108 Coine Road, Twickenham, Middx.

"P.W. Sett., Oct. and Dec. 1970 and Oct., Nov. and Dec. 1971.—C. J. Page, Flat 6, 17 Devonshire Terrace, Lancaster Gate, London, W.2.

"April 1971 Issue of P.W.—T. M. Fresson, Fernilee, Churt, Farnham, Surrey.

"P.W. July 1960.—T. Richards, 52 Cookson Street, Liverpool, L1 5EH.

"P.W. Jugust 1988 and issues containing the Portable Keyless Organ.—K. Oliver, 10 Mayflower Road, Shirley, Southampton, Hants.

"P.T. July 1971 with the 625 i.f. strip.—O. Collins, Landford, Germansweek, Bearworthy, Devon."

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JULIAN ANDERSON

A series of simple transistor projects, each using less than twenty components and costing less than one pound to build.

HIS month's project is one for the DXers; it is a noise limiter circuit that can either be built into a receiver or used as a separate unit, in which case it fits between the headphones and the receiver.

Noise is a real curse when listening to weak, far off stations and of course its amplitude bears no relationship to that of the signal that you want to hear. A comparatively simple circuit, Fig. 1, will eliminate all signals above a certain level while unaffecting those below that level. Our ears have a sort of a.g.c. system built into them, if they are subjected to a series of loud noises they will not be very sensitive to quieter passages in between those bursts, so, however loudly you turn up the volume, this will not improve intelligibility. By limiting the level of the peaks to that of the wanted signal there is a considerable improvement and signals which were quite lost in the noise can be heard comparatively well.

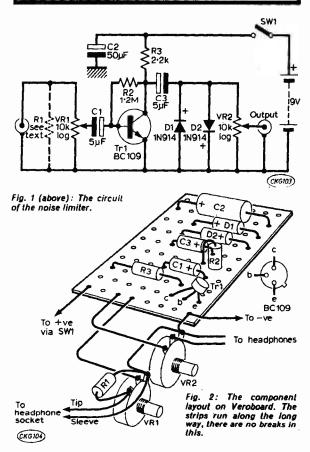
Let us assume that the unit is to be used from the output of the receiver. This output may be the equivalent of an external loudspeaker socket or a special headphone output, it depends on the design and type of receiver. If the socket is for an external loudspeaker and the output stage is a transistorised one, it is important that the output load is maintained and a resistor equivalent to the speaker impedance should be fitted, in the circuit shown as R1. If you are not sure what the value should be 100 will be satisfactory.

The input is connected via the level control VR1 to Tr1 where the signal is amplified. The output of this is fed via C3 to the back-to-back silicon diodes D1 and D2. These are not directly biased but as soon as the level of the signal across these goes over about 0.6V, they will conduct; one of them clips the positive peaks while the other takes care of the negative peaks. It doesn't matter how high the signal level is, anything above this level will appear as a maximum of 0.6V. Since we shall now always have the same level at this point we have removed our control over the listening volume and for this reason VR2 is included.

If the wanted signal level is above the critical 0.6V, this will introduce distortion and it is in this way that the controls are set. The receiver's volume control should be set for normal listening level and, with the unit connected, VR1 should be increased until the wanted signal becomes distorted and harsh; this should then be backed off slightly. VR2 is then set for a comfortable listening level.

The current consumption of the circuit is not high and a small PP3 battery will cope nicely. For fixing the circuit permanently into a set, VR1 can be the existing volume control, it doesn't matter even if the value of this is considerably higher than that

No. 36 Noise Limiter



* components list

Prices are those recently advertised in Practical Wireless and may have changed. No allowance is made for minimum order costs or for postage and packing and these should be checked carefully before ordering.

shown. The output from VR2 can then be connected to the point that was previously fed from the volume control's slider. If used with a valve set the supply voltage can be taken from the h.t. line via a suitable dropping resistor to give 9V. A switch can also be included to switch the unit either in or out.

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quick change Cassette GNOME 150, 150 watt QI with

Cassette LIQUIMATIC, 150 watt QI with 6" wheel

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VELCO ELECTRONICS

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Times in GMT Frequencies in kHz MONTHLY

HE recent spell of cold weather seems to have kept most of the readers of this column in their shacks and at their writing desks, this is evidenced by the number of reports received-the record having been beaten yet again.

DX LISTENERS

The first report comes from Ian Howes in Lowestoft who used his R209 Mk II receiver and TV antenna to log the following interesting stations:-

3375 AIR, Gauhati with Indian Music at 1700.

4780 R. Carabobo, Venezuela, Spanish at 0015.

4860 AIR, Delhi, Domestic Service at 1800.

4880 R. Universo, Venezuela, Spanish at 2315.

4965 SABC, Domestic Service at 2100.

5000 R. Nepal in Nepali at 1540.

5035 Alma Ata, USSR at 1600.

NEWS FOR

5035 Bangui, Cent. Afr. Rep., French at 2230.

5047 R. Lome, Togo, news in French at 2200.

5052 R. Singapore, English Pops at 1600.

9660 ABC, Brisbane, Home Service at 1200.

11835 R. El Espectador, Uruguay, Spanish at 2304.

15430 Voice of Free Korea in English at 1130.

D. A. Hairon of St. Clement, Jersey has again used his Codar CR70A to good effect hearing:-4915 R. Accra, Ghana in English at 2300.

6170 Israel B.C. in English at 2115.

9505 R. Record, Brazil in Portuguese at 0110.

9510 R. Barquisimeto, Venezuela, Spanish at 0035.

9625 BBC, Cyprus relay in Hindi at 0050.

11775 AIR in English at 1050.

11850 R. Accra, Ghana in English at 2030.

11875 R. Soc. de Bahia, Brazil at 0000.

15084 R. Tehran in Farsi at 1430.

15105 BBC, Ascension Is. relay at 0730.

15225 R. Cultural de Bahia, Brazil at 0100.

15230 RT Belgium in English at 0050.

17825 NHK, Japan in English at 0815.

Michael Berry of Dewsbury, Yorkshire is a new reporter to this column, his equipment consists of an Eddystone EB35 receiver and a 15 foot vertical antenna. Amongst the interesting stations heard were:-

4870 Cotonou, Dahomy in French at 2230.

4915 R. Accra, Ghana in vernacular at 2215.

4940 R. Abidjan, Ivory Coast in French at 2305.

6040 Tashkent, USSR in English at 1420.

9520 R. New Zealand in English at 0900.

9570 R. Kaduna, Nigeria in English at 2100.

9570 R. Portales, Chile in Spanish at 0005.

9850 R. Bangladesh, Dacca in English at 1515.

11710 R.A.E., Argentina in English at 2335.

11735 RTV Morocco in French at 1910.

11805 R. Globo, Brazil in Portuguese at 2105.

11865 R. Kinshasa in French at 2230.

11925 Tashkent, USSR in English at 1410.

11925 R. Bandierantes, Brazil in Portuguese, 2145.

15190 ETLF, Addis Ababa in Arabic at 1430.

15520 R. Bangladesh, Dacca in English at 1230.

THE BROADCAST BANDS

Malcolm Connah

Julian V. Moss of Rayleigh used his Meridian 10 transistor superhet and 60 foot long-wire antenna to compile his latest log which includes:-

9515 R. Ankara, Turkey in English at 2200.

9525 AIR, Delhi in English at 1945.

9545 R. Accra, Ghana news in English at 2045.

9605 R. Canada International, English at 2100.

9695 RSA, South Africa in French at 2015.

11620 AIR, Delhi news in English at 1945.

11710 VOA, Liberia sign off at 2230.

11720 R. Canada International, English at 2207.

11780 HCJB, Quito, Ecuador in English at 2000.

11850 R. Accra, Ghana news in English at 2030.

11900 RSA, South Africa in English at 2235.

11910 ETLF, Addis Ababa in English at 2015. 11920 Abidian, Ivory Coast in English at 1845.

11950 FEBA, Seychelles in English at 1730. 11970 RSA, South Africa in English at 2240.

Brian Ewing of Ilford is another reporter with a Codar CR70A receiver, this time used with a 50 foot long-wire antenna to hear:-

6010 RAI, Rome, Italy at 0110.

6155 R. Renascena, Portugal (tentative) at 0030.

9510 BBC, Ascension Is. relay at 0415. 9615 HCJB, Quito, Ecuador at 0330.

9695 RSA, South Africa noted at 2215.

9730 ETLF, Ethiopia at 0400.

11815 TWR, Bonaire noted at 0130. 11850 BBC, Tebrau, Malaysia relay at 2345.

11925 Radio Kuwait at 2040.

11930 WIBS, Windward Islands at 2031.

11955 Radio Lebanon at 0235.

15170 ELWA, Liberia at 2159.

The last reporter for this month is Fred Wall of Walthamstow who has a Plessey PR155 (short pause whilst your scribe goes green with jealousy) and a 20 foot long-wire enabling him to hear: -

9460 Radio Pakistan with music at 2020.

9520 Radio New Zealand in English at 0900.

9745 Radio Baghdad with music at 2040.

9755 Radio New Zealand at 1700.

15240 Radio Australia with news at 2153.

15245 RTV Morocco, news in Spanish at 1415.

15345 Radio Kuwait, Pop Music at 1600.

17820 Radio Ankara, news in English at 1330.

Reports should arrive by the 15th of the month and be addressed to me at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.

THE AMATEUR BANDS David Gibson, G3JDG

Frequencies in kHz • Times in GMT

HINGS are hotting up r.f.-wise at Moretons, Harrow. The "Dorms" are bedecked with sneaky antennas of varying shapes and sizes and are used by an enthusiastic band of s.w.ls. Jonathan Thackeray is one and pleads guilty to listening on the Amateur bands by fixing an H.A.C. one-valver onto his modest 15ft. end fed. If you are having a chuckle at all this, observe Master Thackeray's log for 80 metres; JA5TY, JY1/B, UL71AF, UP2TP, VE1FO, VE3QA, VK2LF, VO1MG, VP9UN, VQ9R, 4X4DP, 9H3B, 9K2AL. Definitely ten out of ten for that splendid effort. (Write out one hundred lines, "I must not receive better DX on eighty than G3JDG!")

It seems that most listeners kept their ears draped over the l.f. bands this past month. Log from **David Lawley** (Gravesend) shows a fair bit of activity on top band. Gear is a CR70A plus 220ft. folded longwire with a quarter wave counterpoise earth for topband. All stations are c.w. unless marked; DJ4SS (s.s.b.), DL1CF, DL9KR, EI8BF, EI9BG, EI9J, GD3GM (s.s.b.), GM3-IGW, UKG, YCB, YOR, ZDH, GW3-UCB, UPK, YTO, HB9-CM, NL, OK1-ATP, AYY, KYS, MA, SBT, OL1AOH, OL5ANJ, PA0PN, W1BB/1, W1HGT. David bagged 195 countries last year, has already heard 85 this year and hopes to take the R.A.E. on May 9 (Gd lk OM).

John Runchman reports "A" level QRM but has managed a few late night/early morning sessions as a nice change from delta y tending to zero! Receiver is a homebrew two-valve t.r.f. (ECC83/ ECL82-what funny transistor numbers) with a 20metre dipole. Topband c.w. brought in: EI8H, HB9ANW, HB9CM, HB9NL, OK1ATP, OK1MAC, PAOPN. Just in case you think that two valves isn't very much you are quite right. The following log received on this set for 14MHz tends to add weight to the argument that it's what's between the headphones that counts; CN8GG, EA6BM/M, EA8HR, G3RFP/MM, LU2AJE, PJ2CL, PY1CBS, PY1DG (no relation!), PY1ZAN, PY2DVH, PY4AP, PY7OS, PZ1DR, UC2DX, VO1CU, VP2MZ, VP9CP, YV4AOW. YV6IP, 4S7SW (Ceylon), 4X4BL, 7X2BK, 9G1FF, 9H1CQ, 9Q5ITU, 9Q5RD. All this with a t.r.f., that should get some "reaction" from the superhet fraternity? Incidentally, John's QTH is South Norwood, London.

News from Karl Muller in the land of 3D6 (that's Swaziland in Southern Africa) that FL8HM and FL8MM both operate on band edge of 7MHz c.w. but he doesn't say the QTR. Karl has a Barlow-Wadley XCR-30 receiver with a 30ft. end fed and an a.t.u. The following made it into Swaziland on 7MHz c.w.; CM3LN, CR6AI, DK1EB, DU1IOR, FB8XX, FL8HM, FL8MM, HC2HM, JA1OHU, JA3EA, JA3PMB/6, JA5ADR, KP4DJE, KR6AY, LA2QI, OK2KYI, PY1BSN, PY3MU, TU2BK, VP2AAA, W6GRU, W7DI, ZD8BR, 4S7DA, 5R8BD, 6W8AL, 9M2CW. Incidentally, on ten metres Karl logged G3UGM.

He also asks if people who work s.s.b. have callsigns and enquires why certain YU and LZ stations spend half an hour calling CQ on top of DX stations? (Yeah, well...).

Patrick Funnell reports on a pirate station on topband signing G4NCC who, apparently, spends his hours taking the mickey out of Amateur radio. Be warned friend, in this transistor age many Amateurs have some very simple but excellent d.f. equipment. All it needs is two of them to hear you and get a fix, and where the lines cross—that's you. Patrick has a BC312 receiver and a "... piece of wire in the loft." This set-up logged these on topband: EI9J, GM3IGW, GM3YOR, GM4AGG, OK1ATP, OK1JAX, OK1KYS, OK1NRX, all on c.w. and heard between 1830kHz and 1840kHz.

Eighty metres seems to be a strange band. Some acclaim it as just great while others give a clear "thumbs down" sign, or something.

Steve Taylor (Abbey Wood, S.E.2) reckons there's a lot of "G" activity about judging by what he heard on his CR7OA. Stefan Kaye (Witney, Oxon) sends in a good log of eighty metre happenings bagged with the aid of an AR88D and 180ft. longwire which is untuned. Pickings include: EA8HA, OA4OS, VE1ADV, VE1AU, VP2LAT, WB0FFG/TF, XE1KB, ZL3GS, ZL3RK, ZL4AV, ZL4KE, 4X4UF, 9H1BX, 9H1C, 9H3B.

"My first log to the Amateur Bands", says John Vaarnela of Croydon Park, New South Wales, Australia. Receiver is a BC348Q and the antenna a dipole at 25ft. Squeaks of s.s.b. on 14MHz heard from: A2CAB, CE1LU, CN2CG, CP1DN, CR6GA, CR7GJ, CX4RZ, EA8GZ, EL0K/MM, ET3JH,, FB8ZZ, FB0WK, FK8ID, FL8MM, FO8DO, FOOTZ, FR7AB, HI8LC, HK1CKY, HK0BKX, HR1BK, HS1ADR, IC8CQS (Capri), JD1KAA, KC6QS, KG6SI, KH6AQ, KR6TQ, KP4DIW. KV4AA. KS6DH. KX6PQ, KZ5GB, LUISE, MP4BBW, OA4PQ, PJ2MI, TI2RT, VP1BH, PY3ATW, VP9AT, VQ9NEW, VR1AB, VR4EE, VR5FX, VS5PW, VS6BS, VS9MT, VU2ED, XE1KB, XW8DO, YA1AG, YB0AT, YJ8BJ, YK1AA, YN1HC, YS1DET, YV6JB, ZD6HJ, ZD8RR, ZE2KV, ZP5KU, ZS2DH, 3B8CJ, 3D2FM, 4S7PB, 4W1AF, 4X4YT, 5H3LV, 5N2ABG, 5R8AB, 5V3RU, 5W1AU, 5X5NA, 5Z4KZ, 9H1ST, 9K2CP, 9M2GV, 9M8FMF, 9N1MM, 9V1VR. Who said twenty was dead?

Julian Iredale (Llandudno), JR500SE, Codar PR30, 264ft. end fed got these goodies on 28MHz s.s.b.: K4AUV, K4HGG, VE40CK, W2BLV, W4GJO, W8QUO, W0ZRX/MM (500 miles north of the Azores), WA5MAK, WB4BMV, WB5CTE, WB4WLI/YV1, ZC4GLW, 4Z4BG, 5R8AP, 7Q7AA, 9J2DG.

Logs, in alphabetical order please, to arrive by the 15th of the month to:

12 Cross Way, Harpenden, Herts.

practically wireless commentary by HENRY

No. 89

Dim out

LAVE to the whims and wishes of my readers—the whims of one and the wishes of the other, did I hear you say?—Henry sits here in the light of a flickering candle, like a Dickensian P.W. Editor.

Henry has been remiss. Not gifted with foresight, he to whom tea-leaves are a greater mystery than Rorsach blots failed to provide auxiliary illumination against the day when the power should be cut. The bulb's gone in me torch, innit?

I should have invented something electronic, on the spur of the moment, say the cold eyes around the blank TV. Like all great discoveries, the answer should have come in a blinding flash—well, perhaps that is not the best simile. . . .

As for example, when the transistor was born? Ah, but I'll bet it wasn't at all like that.

Shockley: Let's give it up. Nearly lunchtime.

lunchtime.

Brattain: One more try. (He's the meticulous one. Like Spencer Tracy as Edison, he will not give up.)

S. I'm hungry. We are not going to get AFN without a tuning coil, and I'm sick of sharing the same bit of germanium.

B. Right. I'll just leave my catswhisker here. You can do what you like with yours.



A Dickensian P.W. Editor

Shockley does.

They return, replete and burping. That Bell Laboratories Inc. canteen was notoriously shepherdpie oriented.

B. Next step, then.

S. Oh., let's start again.

B. No, you measure, I'll inscribe.

S. Just a minute while I suck a—hey—what's this?

B. What's what?

S. Amplification. . . .

B. Well, it is a process of-

S. No, lout. Look! 1mV in, 100mV out. You fiddled the meters.

B. I did not, indeed!

S. Well, somebody did. Let's pack up and invent a fuel cell.

B.—and twenty makes fourhundred—no, listen Shock, we've got something here. We can transfer effective resistance over a link circuit. We can leap a barrier.

S. Don't get romantic. Have some bicarb.

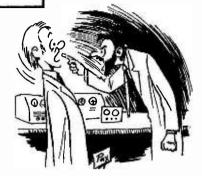
B. Bicarb, hell! Don't you see, this is a breakthrough. No more pottwiddling. Just stick a couple of cats' whiskers on a slab of germanium and the input resistance changes to a completely different output resistance just by an alteration of voltage. . .

Of course, dear reader, it did not turn out like that. Some fussy, officious bookmaker from the front office came down to the bogfins' dimly-lit cellar (that's old Bell Inc. for you—my candle is getting low and smelly). He said: Pack it in chaps. Don't forget we own half the Varires company and control Pots Inc. You can't invent the noiseless pot. Let's put it to another use, or scrap it.

And so the transistor, as an amplifying device, was reluctantly dragged forth.

Well, that's my story. But the truth was probably quite different. Most inventions are the result of sweated persistence. 'Eureka' was probably Greek for 'the water's too hot.'

Henry has seen some of this dogged persistence in action. A great friend is an industrial de-



You fiddled the meters

signer. His development chief presents him with impossible problems a couple of times a week. He is given a pretty free hand—enviable job!— and the limiting factor is the utmost reliability and speed of construction. No jerry-built hook-ups for him. His tobacco-tin wizadry is made of cast-iron.

So I wonder how he is getting along in the dim-out? I only hope he did not try to follow the same circuit my colleague tried with the cadmium sulphide cell. Published, he told me, in Pr*ct*c*l*l*ctr*n*cs (R**lly? Ed.). It did not work, and there was no time to find out why.

So this chap being in his way a latter-day Brattain, persisted in racking his brains for past applications and came up with the trigger circuit for an oven control circuit. Problem—no unijunction transistor. Solution, use a pnp in an involuted hookup—good old Shockley's method.

The final result, with two transistors and four resistors, worked a treat. But as the CDS had to be the lower limiting factor of his resistor network, he now has the problem that there is enough infra-red reflected off the ceiling from a television screen to latch the circuit on.

So what, you say, if there is no power, there will be no telly.

That's where you'd be wrong, you see. The first thing he did was to rig up a battery driven invertor to feed the television receiver!

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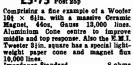
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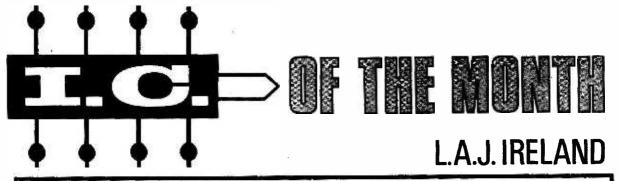
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Number 29

GEC Silicon Switches 2N4990/1

In the past a number of utility type i.c.'s were reviewed in the "I.C. of the Month" series such as the Mullard TAA320 Bifet or the D13V voltage regulator, etc., and it is quite some time now since any of these three or four terminal devices have appeared. Consequently this month's article is devoted to a review of two specialised monolithic i.c.'s namely the Silicon Unilateral and Bilateral Switch (SUS and SBS) which have recently been released by the General Electric Company. Some useful practical applications of the devices are given.

Operation

In appearance the SUS and SBS resemble a typical small signal transistor mounted in an epoxy TO-98 package, the electrical characteristics of the SUS closely resembling those of an ordinary four layer thyristor. The monolithic chip has a pair of complementary transistors, a zener diode and a resistor

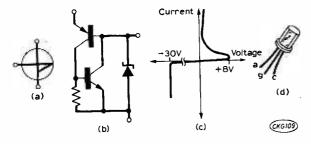
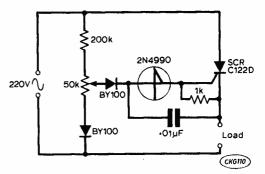


Fig. 1: Circuit symbol (a), equivalent electrical circuit (b), characteristics (c) and outline (d) of the Unilateral Switch 2N4990.

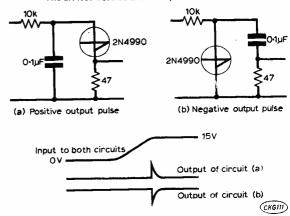
fabricated into it as is shown in Fig. 1. The resulting static electrical characteristics of the device, in which the applied voltage versus current is plotted, is also illustrated and from this graph it can be seen that once the applied voltage exceeds a typical value of about 8 volts rapid forward conduction takes place with the voltage across the device dropping almost to zero.

Applications

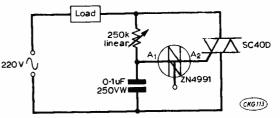
A gate lead is also provided to allow triggering at lower voltages in addition to providing transientfree wave forms, but in most applications this lead



The 2N4990 used as a motor speed control.



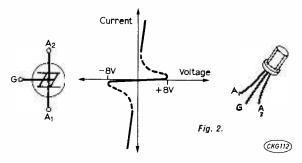
Here the 2N4990 SUS is used to generate a rapid purse from a comparatively slow rising input. Can be used as an over-voltage sensor with the pulse firing a thyristor.



The Silicon Bilateral Switch 2N4991 (SBS) in a mains full wave power control unit.

may be left floating. The SUS therefore is admirably suited for such applications as over-voltage sensing, pulse sharpeners, and SCR triggering. In conjunction with an SCR an efficient and sensitive automatic over-voltage cut-out sensor can be designed very simply and a number of typical applications of the SUS are given.

Just as the Triac can be considered as consisting of a pair of thyristors connected in inverse parallel so also is the SBS related to its counterpart, the SUS. A schematic diagram of the 2N4991 is illustrated in Fig. 2 accompanied by its static electrical characteristics. It fires in both directions so that in opera-



tion it resembles the ST2 Diac which has established itself as one of the most popular triggering devices in Triac power control circuits. However the 2N4991 has tighter specifications producing sharper pulses than the ST2 although it is offered in the same price bracket as the latter.

Both units are available from Jermyn Industries Ltd., Sevenoaks, Kent.

TELEVISION

MAY ISSUE

PAL DECODER

This month we start on the construction of the TELEVISION Colour Receiver and go straight to the heart of things—the PAL-D decoder. This is built on a conveniently sized printed circuit board and uses well tried and tested circuitry. The board can of course be used as a separate module with other equipment if the reader wishes. In building this decoder you will have a real opportunity to learn about colour signal decoding through practical experience.

RENOVATING THE RENTALS

The Pye/Ekco T418/11U series gets our attention this month.

WIDEBAND BAND 1 AERIALS

Essential for DX use but not commonly available is the wideband Band 1 aerial. Along with details of practical designs Roger Bunney also presents information on the fundamentals of wideband television aerial design.

ON SALE APRIL 17



AVID RICHARDSON of West Ealing has been active on the medium waves with his Lafayette HA230 and medium wave loop antenna. He reports Radio Popular La Palmas on 836kHz; CKBW Bridgewater, Nova Scotia on 1000kHz; WHN New York City 1050kHz; CJRP Quebec, broadcasting in French on 1060kHz; WBAL Baltimore 1090kHz; CBI Sydney, Nova Scotia 1140kHz; CKCV Quebec, in French, 1280kHz; WABK Gardiner, Maine also on 1280kHz; WLAM Lewiston, Maine 1470kHz; WCKY Cincinnati, Ohio 1530kHz. David has also logged the Voice of America Okinawa 1178kHz at 1445hrs SINPO 22432, with a programme in Chinese. This outlet is usually heard in Europe during the afternoon in winter. It has a power of 1000kW, it beams westward into China and sometimes competes with Horby, Sweden on the same frequency when conditions are good.

W. Pennington who lives at Grimethorpe near Barnsley in Yorkshire has built the Radio Nederland MW Frame (Loop) aerial. His receiver is a B40D and on Tuesday February 8th he heard CJON St John's on 930kHz, WINS New York 1010kHz; CBA Moncton, New Brunswick 1070kHz. After midnight he logged CBN 640kHz in St John's; CJOX 710kHz Grand Bank, Nfld; Radio Demerara Guyana 760kHz; WHN 1050kHz and WNEW 1130kHz both in New York City. "I thought it was impossible to receive these stations in England but I have proved it for myself and I find it most interesting" he writes and he suggests that readers may like to write to Radio Nederland, Postbus 222, Hilversum, Holland and ask for a copy of The Frame Aerial leaflet, which gives constructional information on a medium wave loop antenna.

Mike Kerry of Golders Green sends an interesting log of Latin Americans heard on his Eddystone 750 and medium wave loop. These include YVQQ Radio Puerto Cruz, Venezuela on 760kHz; Radio Demerara 760kHz located in Georgetown Guyana, with programme in English; R.Oriental, Uruguay 770kHz; La Voz del Rio Cauca, Colombia 820kHz; Radio Mundial, Rio de Janeiro 860kHz; R.Tamandau, Recife 890kHz; Radio Jornal do Brasil 940kHz; Radio Belgrano, Buenos Aires 950kHz; R.Sutatenza, Colombia 960kHz; Radio Record, Sao Paulo 1000kHz; R.Margarita, Venezuela 1020kHz; R.Revolution Cuba 1060kHz; R.Nacional, Sao Paulo 1100kHz; WBMJ San Juan, Puerto Rico 1190kHz; R.Valera, Venezuela 1230kHz; R.Canaima, Ciudad Bolivar, Venezuela 1290kHz; R.Nacional, Santiago in the Dominican Republic 1380kHz. Mike also mentions the VOA Marathon Key, Florida on 1180kHz which broadcasts in Spanish. Latin Americans become prominent as the year advances and it is worth while listening between 0100hrs GMT and sunrise for the stations listed above.

Please send reports and information about the medium waves to the author at 132 Segars Lane, Southport, PR8 3JG.

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2N1131 2N1132	25p 25p	2N3854A 2N3855	271p 271p	AC154 AC176	221p 25p	BDY11 BDY17	£1.621 £1.50	BSY82 BSY90		NKT801	12 971p
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2N1307 2N1308	25p 80p	2N3859 2N3859A		ACY20 ACY21	25p 25p	BDY61 BDY62	£1.25 £1.00	C425 C426	55p 40n	NKT802	92≟p 13
2N1309 2N1507	80p 171p 25p	2N3860 2N3866	£1.50	ACY22 ACY28	20p 20p	BF115 BF117	25p 471p	C428 C744	374p 30p	NKT802	921p
2N1613 2N1631	85p	2N3877 2N3877A	40p 40p	ACY40 ACY41	20p 25p	BF163 BF167	87 p 18p	D16P1 D16P2	87 p 40 p	NKT802	92 է ր 15
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2N1893 2N2147	871p 821p	2N3096 2N4058	871p 871p	AF106 AF114	42 ł p	BF184	0K	GET118 GET119	20p 20p	OC24	- 60a I
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2N2483 2N2484	27 + p	2N5030 2N5172	421n	AF211 ABY26	821p	BFX44 BFX68	87 i p 67 i p	MJ1800 MJE340	62-17‡ 62-1	OC81 OC81D	20p 221p
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2N2613 2N2614	85p 80n	2N5176 2N5232A	45p 30p	ASY29 ASY36	27½p 25n	BFX86 BFX87	25p 271p	MPF102 MPF103	421p 871p	OC139 OC140	821p 821p
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2N2904A 2N2905	821p 871p	2N5307 2N5308	871p 871p	BC109 BC113	10p	BFY20 BFY21	£1.60 42 p	NKT135	27±p 27±p	OC207 OCP71	75p
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2N2906A	271p	2N5354	42 p 27 p	BC116A BC118	15p 10p	BFY26	25p 20p	NKT211 NKT212	80p 80p	ORP61 P346A	50p 221p 621p
2N2907 2N2923	80p 15p	2N5355 2N5356	271p 821p 471p	BC121 BC122	20p 20p	BFY29 BFY30	50p	NKT213 NKT214	221 p	TIS34 TIS43	27p
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2N3055 2N3133	62р 80р	28502 28503	85p 27∤p 40p	BC160 BC167	621p	BFY90 BFW58	421p 571p 671p 271p	NKT240 NKT241	271p	TIS60 TIS61	25p
2N3134 2N3135	80p 95n	3N83 3N128	70n	BC168B BC168C	10p 11p	BFW59 BFW60	25p	NKT242 NKT243	20p 621p	TIS62 TIP29A	271p 50p
2N3136 2N3390	25p 25p	3N140 3N141	77 i p 72 i p	BC169B BC169C	11p	BPX 25 BPX 29	£1.65 £1.80	NKT244 NKT245	174p	T1P30A T1P31A	60թ 62 եթ
2N3391 2N3391A	zup	3N142 3N143	55p	BC170 BC171	12 lp 15p	BPY10 BRY39	\$1.45	NKT261 NKT262	20p	TIP32A TIP33A	75p
2N3392 2J3393	17ip	3N152 R.C.A.	67 lp 87 lp 52 lp	BC172 BC175	15ր 22էր	BSX19 BSX20	871p 171p 171p	NKT264 NKT271	. 20p		1.021p £2.05
2N3394 2N3402	15p 224p	40050 40251	521p 55p 821p	BC182 BC183	10p 09p	BSX21 BSX26	87≩p 45p	NKT272 NKT274	20p	TIP35A TIP36A	£2.90 £3.68
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SN7430	0.23	0.15	SN7460	0.20	0.18	8N7495	0.87	0.8
SN7432	0.48	0.42	8N7470	0.40	0.38	SN7496	0.87	0.8

USES range axial lead Values: (\(\mu F/V\); 0-64/64: 1/40; 1-6/25; 2-5/16; 2-5/64; 4/10; 4/40; 6-4/6-4; 6-4/25; 8/40; 10/16; 10/84; 12-5/25; 16/40; 20/16; 20/64; 25/8-4; 25/25; 32/10; 32/40; 32/64; 40/16; 50/6-4; 50/25; 50/40; 64/10; 80/2-5; 80/16; 80/25; 100/6-4; 126/10; 125/16; 200/10.

	S	ILIC	ON	REC'	TIFIE	RS		
PIV	50	100	200	400	600	800	1000	1200
1.A	,8p	. 9p	10p	11p	19p	15p	20p	
3A	15p	17p	20p	221p	25p	27p	80p	85 p
6A 10A		F01 -	25p	80p	82 i p	85p		
15A		521p	571p	65p	771p	861 p	971p	£1 · 25
35 A	_	57∳p 80p	62 j p 90 p	771p	90p £1·25	971p	\$1.20	\$1.571
lamp and a						£1·50	£2·50	

		IERS	RECTIF	ES &	DIODE		
82 1	F8T3/4	124p	BAX16	7p	AA119	10p	IN34A
175	OA5	17 tp	BAY18	15p	AA129	70	IN914
g05	OA10	7p	BAY31	12p	AAZ13	7p	IN 916
ĹΟ̈́ρ	OA9	25p	BAY38	12p	AAZ15	20p	IN 4007
8p	OA47	15p	BY100	10p	AAZ17	7p	IS44
75	OA70	22p	B¥103	15p	BA100	15p	IS113
105	OA78	471p	BY122	25p	BA102	12p	18120
79	OA79	15p	BY124	25p	BA110	14p	18121
8p	OA81	15p	BY126	15p	BA114	8p	18130
10p	OA82	17p	BY127	7p	BA115	10p	18131
7p	OA90	57p	BY164	17p	BA141	12p	18132
79	OA91	22p		17p	BA142	7p	
79	OA95	85p	BYZ10	120	BA144	8p	18922
79	OA200	82p	BYZ11	17p	BA145	12p	18923
10p	OA202	80p		12p		5p	18940
	OA9 OA47 OA70 OA78 OA79 OA81 OA82 OA90 OA91	25p 15p 22p 47ip 15p 15p 17p 57p 22p 85p	BAY38 BY100 BY103 BY122 BY124 BY126 BY127 BY164 BYX10 BYZ10	129 109 159 259 259 159 70 179 179 129	AAZ15 AAZ17 BA100 BA102 BA110 BA114 BA115 BA141 BA142 BA144	20p 7p 15p 12p 14p 8p 10p 12p 7p 8p	IN 4007 IS44 IS113 IS120 IS121 IS130 IS131 IS132 IS920 IS922

OPT	OPTOFI ECTPONICS PRINCE PECTIFICE							
		BAX13	5p	BYZ13	25p	TIV307	50p	
18940	5p	BA154	12p	BYZ12	80p	OA202	10p	
18923	12p	BA145	17p	BYZ11	82p	OA200	7₽	
18922	8p	BA144	12p	BYZ10	85p	OA95	79	
18920	7p	BA142	17p	BYX10	22p	OA91	79	
18132	12p	BA141	17p	BY164	57p	OA90	7p	
18131	10p	BA115	7p	BY127	17p	OA82	10p	
18130	8p	BA114	15p	BY126	15p	OA81	8p	
IS121	14p	BA110	25p	BY124	15p	OA79	70	
18120	12p	BA102	25p	BY122	471p	OA78	10p	
IS113	15p	BA100	15p	BY103	22p	QA70	7p	
I844	.7p	AAZ17	10p	BY100	15p	OA47	8p	
IN 4007	20p	AAZ15	12p	BAY38	25p	OA9	10p	
IN 916	7p	AAZ13	12p	BAY31	7p	OA10	20p	
111014		AAILO	100	DAIIO	1/40	UAU	1/9	

OPT	OELE	CTRON	ICS	BRI	DGE R	ECTIFIER	RS
		BAX13	5p	BYZ13	25p	TIV307	50p
18940	5p	BA154	12p	BYZ12	80p	OA202	100
18923	12p	BA145	17p	BYZ11	82D	OA200	79
18922	8 ₀	BA144	12p	BYZ10	85p	OA95	79
IS920	7p	BA142	17p	BYX10	22p	OA91	7p
18132	12p	BA141	17p	BY164	57p	OA90	7p
18131	10p	BA115	7p	BY127	17p	OA82	10p
18130	8p	BA114	15p	BY126	15p	OA81	8p
I8121	14p	BA110	25p	BY124	15p	OA79	7p 8p
18120	12p	BA102	25p	BY122	471p	OA78	10p

OF I GELEC I RONICS	עואם	OE K	ECII	FIEL	12
MINITRON 3015F 7-SEGMENT INDICATOR (16 PIN DIL) 28-00 May be driven by 8N7447 GNP-7AH COLD CATHODE TUBE SIDE VIEWING. 0-9 and TWO DECIMAL POINTS. May be	1 4 140 2 50 2 200	37p 57p 32p 41p 46p	A. 4 4 4 6 6 6	PIV 50 100 400 50 200 400	6 7 8 6 8
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32 × 5in 5 × 17in (Plain)	80p 82p	29p				
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	oole (·1	and -18				

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1 watt	2%, M/O	4p
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2 watt	10%, 6p	E24 Series.

MULL			M/F	OIL
2 400	46p	6	400	£1·10
2 200	41p	6	50 200	62p 80p
2 50	82p	4	400	80p
1 100 1·4 140	87p 57p	4	100	60p 70p

CAPA	CIT	ORS		
0.01, 0.0	22,	0.033,	0.047	Sp each
0.068, 0.				4p each
0.15, 0.2	2, 0.2	33		5p each
0.47				9p
0.68				11p
1μ F				14p
1.5µF	• •			21p
2.2µF	•			25p

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GOING BACK... 1970 60 50 40 30 20 COLIN RICHES ARTHUR DOW

T was on Wednesday, April 10th, sixty years ago that the s.s. *Titanic* left Southampton on her maiden voyage. Five days later, she sank to the bed of the ocean—a useless mass of twisted and torn steel, for on Sunday, April 14th, 1912, at about 11.40 p.m. the greatest peacetime disaster in maritime history took place when this ship, gross tonnage 46,328, struck an iceberg, 50-100ft. high, while on her maiden voyage to New York.

The blow on the starboard side of the ship ripped a hole 300 feet long and made many of the essential watertight compartments useless. She foundered within three hours, in water two miles deep.

Mr. H. S. Bride, the operator who survived the disaster described the way in which wireless played its all important role in saving those who were rescued. Mr. Bride was the second operator on the *Titanic* and was relieving his chief, Mr. Phillips at the time when the collision occurred. "I was standing by Mr. Phillips telling him to go to bed when



Jack G. Phillips, Chief Marconi Operator on board the Titanic.

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(Fitted with Marconi Wireless Apparatus.)
"OLYMPIC" sails from Southampton and Cherbourg to New York regularly.
"TITANIC" sails from Southampton and Cherbourg on first voyage to New York

April 10, 1912

LIVERPOOL, LONDON, SOUTHAMPTON: NEW YORK

How the maiden voyage of the Titanic was advertised.

the captain put his head into the cabin. 'We have struck an iceberg!' he said, 'you had better get ready to send out a call for assistance. 'Don't send it until I tell you'. The captain went away but in ten minutes came back. We could hear terrible confusion outside but not the least thing to indicate any trouble. The wireless was working perfectly. 'Send a call for assistance' ordered the captain. 'What call shall I send?' Phillips asked. 'The regulation international call for help—just that' was the reply, and Phillips began to send the signal C.Q.D. (now replaced by S.O.S.). After a few minutes the Captain reappeared and said, 'Send S.O.S.—it may be your last chance'."

The Titanic's C.Q.D.'s and S.O.S.'s were first picked up by the German steamer, Frankfurt which was 153 miles away. Almost at the same time, the Carpathia's wireless operator reported the emergency to his captain and was able to give the stricken liner's position 41 46N 50 14W. Immediately, the Carpathia, which was 58 miles away, altered course to the rescue. The first message the Carpathia picked up was, "COME AT ONCE, WE'VE STRUCK A BERG. IT'S C.Q.D. OM C.Q.D." C.Q.D. of course meant Come Quick, Danger, and was rather tricky to tap out in Morse. For this reason, the now universal distress signal S.O.S. was also used and this was the first occasion it had really been broadcast in earnest.

Fortunately, the Carpathia's only wireless operator, although officially off watch had returned to

his equipment and was putting out some routine traffic calls—including some to the *Titanic*. At 12.20 a.m. the distress calls were received and soon at least six ships were steaming to the disaster zone.

At 2.20 a.m. the Californian which for sometime past had been watching the lights of the 'unknown' vessel grow steadily dimmer, noted that they had finally disappeared. The optical illusion being that of seeing a ship disappearing into the darknessnot going under the waves. Those witnessing the scene from the decks of the Californian were not to know that they had been watching the death throes of the Titanic and that the sea around them was dotted with hundreds of human beings fighting for their lives. Two long hours were to elapse before the Carpathia arrived and began to pick up survivors. Among over 1,500 people lost on that fateful night was Jack Phillips, the *Titanic's* senior wireless operator, who remained at his post as the decks were awash. Mr. Bride, who showed an equal devotion to duty, was eventually rescued after nearly two hours in the sea.

As the decks were awash, Phillips was standing in the wireless room, still sending away giving the Carpathia details of how the *Titanic* was faring. He then picked up the *Olympic* and told her that *Titanic* was sinking by the head. As Phillips was sending the message, Harold Bride strapped his lifebelt to the wireless operator's back. He had already put on his overcoat and was wondering if he could get Phillips' boots on him while he was still sending.

The Captain's voice boomed, "Men, you have done your duty. You can do no more. Abandon your cabins now. It's every man for himself. You look out for yourselves. I release you—that's the way of it at this kind of time, every man for himself."

The boat deck was awash. Phillips clung on sending...sending. He clung for about fifteen minutes after the Captain released him—water washing the floor of the wireless cabin.



H. S. Bride, 2nd Operator on board the Titanic.



Cyril Evans, Marconi Operator on board the Californian.

From aft came the strains of music—the ship's band playing ragtime. Phillips ran aft and made his way to the deck. He swam to a life-raft but lay there exhausted until his last breath failed. Harold Bride later said of Phillips. "He was a brave man and stuck to his key until the very end. If he had had a chance to go to his room and get warmer clothing, as I did, he would probably be alive today. But duty was first with him."

When water flooded the upper decks, Second Officer Lightoller was up to his ankles in water and Captain Smith—a calm grave figure of a man realising that the end could not now be far off—stood calmly on the bridge. He spoke two words to the people who were crowding forward. Words which were a call to our traditions of race and manhood, "Be British!"

Besides calling the Carpathia and Baltic, Phillips had also contacted the Frankfurt again and vessels had also been sent to the rescue by the wireless station at Cape Race. They were however all very far away and even though they made all speed to the stricken Titanic, not one of them was near when the last terrible moment came.

The band still played on. They played until they were waist deep in the cold dark water and in that awful moment, W. Hartley, the conductor spoke to his colleagues and there rang out into the darkness the hymn NEARER MY GOD TO THEE.

Part 2 of the Titanic story will be published in "Going Back" next month.

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100μA £3.45 100-0-100μA £3.85 500μA £3.20

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100 x 80	mm.
20V. D.C	£3·10
50V. D.C	. £3·10
300V, D.C.	£3-10
1 amp. D.C.	£3·10
5 amp D.C.	£3·10
300 V. A.C.	
VII Motor	49.75

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	. £2 ·60
VII Motor	20 OF

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Type MR.85P. 41in. × 41in. fronts.

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50-0-50μA £2-60	20V. D.C £2.0
100µA £2.60	50V. D.C 22-0
100-0-100µA £2-50	
500μA £2-30	
lmA £2.00	
5mA £2.00	
10mA £2.00	
50mA \$2.00	

50-0-50μA £2·60	20V. D.C £2.00
100µA £2·60	50V. D.C £2.00
100-0-100µA \$2.50	300V. D.C. 42.00
500μA £2·80	15V. A.C £2:10
1mA £2·00	300V. A.C \$2.10
5mA £2:00	8 Meter 1mA 22:10
10mA £2:00	VU Meter \$3.20
50mA £2.00	1 amp. A.C. 22.00
100mA \$2.00	5 amp. A.C. \$2.00
500mA £2.00	10 amp. A.C. # #2-00
1 amp £2.00	20 amp. A.C.* \$2.00
5 amp £2.00	30 amp. A.C.* £2.00
	in.×8iin. fronts.
50μA £3·37	10V. D.C 22 ·20

Туре	MR.	65P.	3 in	.×8	in.	front	8.
μΑ		£3.3	7	10 V.	D.C		22

50-0-50μA £2.75	20V. D.C £2.20
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5mA £2.20	300V. A.C £2:30
10mA £2.20	500V. A.C £2:30
50mA £2.20	8 Meter 1mA £2.37
100mA £2:20	VU Meter 23.37
500mA £2:20	50mA A.C. 42.20
1 amp £2-20	
	100mA A.C.* \$2-20
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5Ad.c 84 40	5V/50V d.c. 24.65
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Type MR.38P. 1 21/32in. square fronts.

	PROPERTY AND A PROPERTY	200mA	21.AC
	100 - 200	300mA	
	Similari de la companya de la compan	500mA	
		750mA	
	, <u>, , , , , , , , , , , , , , , , , , </u>		
i	are are	1 amp	
		2 amp	
	#	5 amp	
	TO1	10 amp	
	50μA £2·10		B1·60
1	50-0-50μA £1.90		e1 · 60
1	100μA £1 90		E1-60
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	200μA £1.75	100V. D.C.	1.60
	500μA £1 65	150V. D.C.	
	500-0-500μA £1-60	300V. D.C.	1.60
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	1-0-1mA £1-60	750V. D.C. 4	1.60
	2mA £1.60	15V. A.C 4	
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150mA	£1·60	VU Meter .	. £2·1
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lmA.	£1.70		
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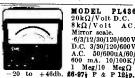
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2N718A 309 2N3691 159 40309	32p BC157 15p BFY30 40p NKT401 87p		6AG7 40p 35Z5 50p PC88 60p
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BSR MP60	£10.00
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Goldring GL72/P	£29 25
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Thorens TD125AB	£90 00
Thorens TDI50 Mk. II	£28 · 00
Thorens TD I 50A Mk. II	£33 · 75
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AMPLIFIERS	3
Please add 75p P. & P.	
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Armstrong 521 (teak cased)	£43 · 95
Alpha Highgate 212	£25 · 00
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Cambridge P40	£65·95
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(Wood cased)	£47 · 50
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(Metal cased) Leak Delta 30	£48 · 00
Leak Delta 30	£56 · 00
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Metrosound ST60	£47 25
Pioneer SA600	€58 00
Pioneer SA700	£66 · 50
Pioneer SA800	£73 · 95
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Sinclair PRO60 2 x Z30/PZ5	£15·25
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Lux 503	£74 .00
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	TUNERS	

Please add 75n P & P

Armstrong 253	£39 - 50
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Rogers Ravensbrook FET4 (Chassis)	£31 -00
Rogers Ravensbrook FET4 (Cased)	£35 · 00
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Leak Delta AM/FM (Cased)	£66 · 50

TUNER/AMPLIFIERS Please add 75p for P. & P.

Alpha Highgate 150	£44 · 95	i
Alpha Highgate R500	£64 · 50	
Armstrong 525 (Teak cases	d) £68 · 50	ı
Armstrong 526 AM/FM		
(Teak cased)	£78 · 50	۱
Leak Delta 75	£130 . 95	
Philips RH781	£50 00	ı
Philips RH 882 (+ cass hea	d) £72.00	ı
Philips RH702	€82 - 50	ı
Teleton 2100	€29 95	i
Goodmans One Ten	£105 · 00	,
Rogers R/brook (Teak)	€80.00	١
Rogers R/brook (Chassis)	£74.50	,

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RADIAC SURVEY METERS—Portable unit as 3 ranges 3, 30 & 300 Rongt per hour with carrying case and inst. battery operated gen on batteries, supplied new cond. £3.00. VALVES 6SNT 5 for 50p. OB2 stab new, 2 for 50p. 85A2 stab new, 2 for 50p. UNISELECTORS Coil 600 ohm contacts 1p 11w 3b small type 80p. TRANS Pria 230v Secs 6.3v 1.7a & 6.3v .3a both ct small C core 50p., also 6.3 x 3 times at 8, 2 & 1a 80p. AIRCRAFT UNIT contains 4 type V60/20p power transis, 2 24v relays, 3 toroid trans, height det unit with magnetic pick-off all in useful case, size 8 x 8 x 4" £1:80. MORSE KEYS fully enclosed key on base 2 x 5" adjustable with Send/Rx swt cord & plug. American surplus £1:10. SPEAKER UNITS P.A. type by Truvox in sloping front case size 9 x 9 x 8" 8 ohm spk with 600 ohm line trans. New boxed £2.75. BATTERIES lead acid type 6v 40 A/Hr in metal case with carrying handle and lid. New boxed with inst. £2.75. Meter Unit cross pointer ind two 115 μ A movements £1.40. O/P & T.U. contain 2 x LA2503 Vinkors, 4x transis 4x Min relays etc. New 85p. FiLTER UNIT contains 10 type LA.2 pot cores, 10x 1% conds & 10 wire ended neons £1-15, METER UNITS radio altimeters F.S.D. 6-5 Ma size 32 sq. 270° 95p. H/Phone Lead with standard jack plug 5ft fitted high to low impedance matching trans new 50p. DIODES 800 P.I.V. at 1a new 12 for £1 .96. TRANSISTORS type OC42 P.N.P. swt type long leads on boards 12 for 50p. RELAYS 12v 180 ohm 4p c/o contacts m.b.b. 50p. AMP Unit small amp 115v 50 c/s I/P valves 6x4, 12Ax7, 6AK6 500 ohm o/p with circ & mods fully enclosed. American sur-500 oftm o/p with circ & mods fully enclosed. A merican surplus £1-80. U.H.F. Rx covers 330 to 335 Mc/s 24v 1/P requires ext. L.O part of landing aid £3-06. FiLTER UNIT contains 20 type LA.2 pot cores & 20 1% conds. all in screened box £1-45. CONTROL BOX contains 4x log swts, yax swt pots etc. 55p. CONTROL UNIT contains valves 6BA6, 12AU7, 6AS6 large pot cores size 1½ x ¾" luf 1% cond, 1x Relay 2 x 8uf 350v PTFE y/h & screens. New boxed 95p. CABLE 6ft length, of 5 core cable, 3 for 56p. RADIO MAGS American, 73 mag new back issues 8 for £1 15. CONTROL UNIT with 2x 120v 5a P.N.P. transis, 2x 50 ohm relays, 3x 6.8v 1w zeners etc. £1.00. PLUG & SK Plessey type 6 way with cable clamp ex new equip 30p per pair. INVERTORS 28v to 115v 400 c/s 180 Va new by Cossor £12.50. E.H.T. COND 1x. 5uf 2 kv wk size 4 x 2 x 2" inc term 50p. TAG STRIPS solder type 24 way, New boxed 10 for 25p. CIRCUIT DIAGRAMS Rx.s 1155B & N.1475, 1392 all 30p. SCOPES CD518 (CT316), 13A, 1035Mk1/2/3, 1049 Mk 3 all 50p. CD513.2 £1-50. CT38 50p. W.M. Class D No. 2 30p. SIG GEN TF801a 75p. AMP UNIT with meter 100-0-100 Us 14" sq. 100 ohm Helipot, 20 pos atten swt. EF86 and ECC83 valves, yax swt, pots etc. on frame 7 x 6" £2:00. TEST SKS. small coloured 12 for 25p. TOROID TRANS 400c/s 2½" dia core 50p. TUNING CONDS. 4 gang 230pf per section air spaced with r.h. drive £1:30. HOUR METERS. 115v 50c/s reads to 99.999 hrs. 60p. C.R.T. type 12SP7 12" Radar Type, new, £3:50. PLUGS & SKS miniature type 14 & 18 way, both 15p pair. TUNING COILS. Tx type 41 x 21 all wound for S.W. new, 2 for 40p.

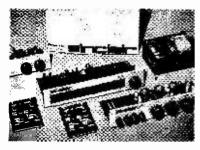
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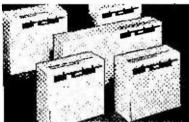
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Sinclair Project 60

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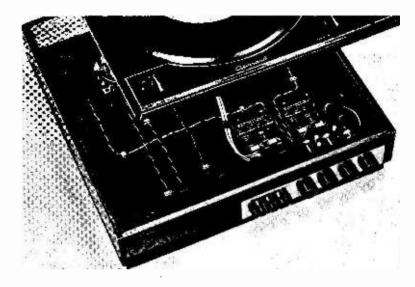
Project 605 is one pack containing: one PZ5, two Z30's, one Stereo 60 and one Masterlink. This new module contains all the input sockets and output components needed together with all necessary leads cut to length and fitted with neat little clips to plug straight on to the modules. Thus all soldering and hunting for the odd part is eliminated. You will be able to add further Project 60 modules as they become available adapted to the Project 605 method of connecting.

Complete Project 605 pack with comprehensive manual, post free £29.95

All you need for a superb 30 watt high fidelity stereo amplifier.

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Project 60 offers more advantage to the constructor and user of high fidelity equipment than any other system in the world.

Performance characteristics are so good they hold their own with any other available system irrespective of price or size.

Project 60 modules are more versatile — using them you can have anything from a simple record player or car radio amplifier to a sophisticated and powerful stereo tuner-amplifier. Either power amplifier can be used in a wide variety of applications as well as high fidelity. The Stereo 60 pre-amplifier control unit may also be used with any other power amplifier system, as can the AFU filter unit. The stereo FM tuner operates on the unique phase lock loop principle to provide the best ever standards of sensitivity and audio quality. Project 60 modules are very easily connected together by following the 48 page manual supplied free with all Project 60 equipment. The modules are great space savers too and are sold individually boxed in distinctive white and black cartons. With all these wonderful advantages, there remains the most attractive of all — price. When you choose Project 60 you know you are going to get the best high fidelity in the world, yet thanks to Sinclair's vast manufacturing resources (the largest in Europe) prices are fantastically low and everything you buy is covered by the famous Sinclair guarantee of reliability and satisfaction.

Typical Project 60 applications

System	System The Units to use together with		Cost of Unit
		Crystal P.U., 12V battery volume control	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U. volume control etc.	£9.45
20 + 20 W. stereo amplifier for most needs	2 x Z.30s, Stereo 60, PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
20 + 20 W. stereo amplifier with high performance spkrs.	2 x Z.30s, Stereo 60, PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc.	£26.90
40 + 40 W. R.M.S. de-luxe stereo amplifier	2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43

Project 60 Stereo F.M. Tuner

Built and tested, Post free.



The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz. Capture ratio: 1.5dB. Sensitivity: $7\mu V$ for lock-in over full deviation. Squelch level: $20\mu V$. Signal to noise ratio: >65dB. Audio frequency response: 10 Hz -15 KHz $(\pm 1dB)$. Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: $2\mu V$. Cross talk: 40dB. Output voltage: $2\times 150mV$ R.M.S. Operating voltage: 25-30 VDC.

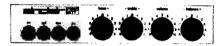
Indicators: Stereo on ; tuning. Size: 93 x 40 x 207mm,

Stereo 60 Pre-amp/control unit

Built, tested and guaranteed.

£9.98

£25



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

SPECIFICATIONS—Input sensitivities: Radio — up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A curve ±1dB:20 to 25,000 Hz. Ceramic p.u. — up to 3mV: Aux — up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE + 12 to — 12dB at10 KHz: BASS + 12 to — 12dB at10 Mz. Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm.

A.F.U. High & Low Pass Filter Unit

Built tested and guaranteed.

£5.98



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages – rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V, Current – 3mA. H.F. cut-off (—3dB) variable from 28kHz to 5kHz. L.F. cut-off (—3dB) variable from 25Hz to 100Hz. Distortion at 1kHz (35V. supply) 0.02% atrated output. Size: 66 x 40 x 90 mm.

Z.30 & Z.50 power amplifiers

Built, tested and guaranteed with circuits and instructions manual. **Z.30 £4.48 Z.50 £5.48**



The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at 15w (8 $_{\Omega}$) and all lower outputs. Whether you

use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.

Input sensitivity: 250mV into 100 Kohms (for 15w

SPECIFICATIONS (Z.50 units are interchangeable with Z.30s in all applications).

Power Outputs

Distortion: 0.02% into 8 ohms.

2.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts.

2.50 40 watts R.M.S. into 3 ohms using 40 volts:

Distortion: 0.02% into 8 ohms.
Signal to noise ratio: better than 70dB unweighted.

30 watts R.M.S. into 8 ohms using 50 volts.

Frequency response: 30 to 300,000Hz±1dB.

For speakers from 3 to 15 ohms impedance. Size: 14 x 80 x 57 mm.

Power Supply Units



Designed special for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 where a stabilised supply is essential.

PZ.5 30 volts unstabilised £4.98 PZ.6 35 volts stabilised £7.96 PZ.8 45 volts stabilised (less mains transformer) £7.98 PZ.8 mains transformer £5.98

Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.



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Special versions of our P50 Series are now available for AF117 or OC45 Transistors. They can be used in the standard superhet circuit with slight changes in component values.

Oscillator Coil	P50/1AC (For OC45)	P50/1AC	(For A	F117)	33p
1st l.F. I ransformer	P50/2CC (For OC45)	P51/1	(For A		36p
2nd I.F. Transformer	P50/2CC (I	For OC45)	P51/2	ÌFor Α		36p
3rd I.F. Transformer	P50/3CC (For OC45)	P50/3V	For A	F117)	36n

Rod Aerial	RA2W	72n
Driver Transformer	LFDT4/1	58p
Output Transforme	rOPT1	58n
Printed Circuit	PCA1	58r

I.F. TRANSFORMERS FOR "PRACTICAL WIRELESS" CIRCUITS

Components for several receivers are available, including the following for the "Clubman".

T41/1E	1st I.F. Transformer	39p
T41/2E	2nd I.F. Transformer	39p
T41/3T	3rd I.F. Transformer	57p
T41/3T	B.F.O. Coil	57p

Details of these and our other components are given in an illustrated folder which will be supplied on request with postage please.

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Amplifier Unit Type A413. Complete with 5Z4 & 6V6GT valves. 250/230 Volt Transformer. Clean condition untested. £1:25. Carriage 75p.

Air Position Indicator. Contains Gears. Ball Races. Dimmer Unit. Lamps. Four numbered Counters. 24 Volt Motor etc. unit of great interest to model makers. Good condition 75p. Carriage 50p.

Ground Position Indicator. Contains 24 Volt Lamp. Prisms Mirror and Gears. Could Interest schools and model makers. 75p. Carriage 75p.

Latest Air Ministry Release Radio Receiver. Type R4187. 2-8 to 18 MHZ In Three Bands. Complete with 17 miniature valves. 26 Volt Motor. Power supply metor and manually tuned. Very clean, but not tested. £5. Carriage and Insurance

Control Unit. Type 4190. Complete with three miniature valves. $1\frac{1}{2}$ ° 500 μ A Meter. Six Relays. Model makers Motor with Gears. 24 Press Buttons. Lots of spares. Clean condition. £1-50. Carriage and Insurance 75p.

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30FL14	-67	DY87	-23	EF86	-28	PCC89	43	PY81	-23	Z77	∙18
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Left and right push-builton on/olf swritches for speakers. Noise intering 20-and lape monitoring facilities. Two auxiliary AC outlets. Frequency response 20-20,000 Hz + 1 db at full power. 15 wats rms per channel. Walnut cabinet with satin aluminium trims. Inputs phono 2-5mV and 5mV RIAA; tuner/aux 250mV. Hum and noise: phono - 50db; tuner/aux - 65db. How's that for a socification! Size 14½" wide. 3½" high, 10½" deep

OLSON RA-310 AM/FM/MPX

This ROC Tuner is especially de signed to match the Dison AM 39! Stereo Amplifier, in price and value, as well as it's good look-

As reviewed in

February 1972

POPULAR HI-H"

value, as well as it's good look-ing design! But of course it's also ideal for use with any other amplifier. The RA-310 costs £10-00 less than the normal retail value, and yet it is a highly sophisticated unit, incorpo-rating the latest solid state techniques. Operation is drift trea for suprating the latest solid state techniques. Operation is drift free lor sup-com³ Price (50 00) reme station-holding capability. You can connect this Tuner to a stereo amplifier, to a tape deck or a tape recorder, And of course it covers all FM Sensitivities: FM, 3ptv. AM, 25ptv. Stereo separation 30dB at 1kHz, image rejection 80dB. Size: 11 % wide, 4" high, 7½" deep.

REALISTIC SA-100B 6-WATT

STERED AMPLIFEER

Here's fabulous, exciting value in miniature! This high quality stered amplifier measures only 9" wide x 3" high :: 52" deep. And yet it has sepa-

rate ganged volume, balance and tone con-trols. Plus speaker in/out, mono/stereo, phono/ tuner and power on/off slide switches. The ends are oiled walnut, with match-

£14.50 ing enamelled metal top. The front namel is satin aluminium and walnut-brown enamel. Frequency response is 50 to 10,000 Hz ± 3dB. Output 3 watts r.m.s. al Price £21-00 per channel into 8 ohms. Inputs are 100mV for both phono and tuner.

REALISTIC TM-100 AM/FM / MPY

ou fabulous value in miniature

the Healistic SA-Tubu in both appearance, size and performance, the TM-100 is superb value-for-money, it gives you the full FM and AM ranges – FM. 88-108 MHz; AM. 535-1605 kHz, Sensitivities AFMSpV; AM 250µV, Image rejection 50dB.

R.446 3-WAY MATCHED SPEAKERS

These will do justice to your amplifier – and to your pocket. At only £16-40 a pair, they are real value-for-money Each cabinet is heavily lagged and teak finished. They handle 16 watts rms (8 watts rms each). Each inflated the handle 16 watts rms (8 watts rms each). Each loudspeaker contains a large dual cone base unit, plus a separate tweeter, Frequency range; 40 to 19,000 Hz. Size 14" high, 9" wide 61" deep

OLSON AM-357 4-WATT STEREO AMPLIFIER

Here's marvellous value for someone just starting to set them selves up in audio! At only £10-50, you get a fine amplifier in scratch resistant metal cabinet, with a smart brushed aluminius front panel. It incorporates separate tone and volume controls for each channel. Inputs are provided for turntable (ceramic cartridge), tuner and tape deck or recorder. Frequency response : 70-20,000 Hz + 3dB. Dutput : 2 watts r.m.s. per channel into 8 ohms. Inputs: phono 80mV; tuner/aux 80mV. Size 8" wide, 21" high

RFALISTIC HI-FI 36-WATT STEREO CONCERTMASTER FM.AM RECEIVER SYSTEM, MODEL 12-694 Brilliantly designed tuner/amplifier with two matching speakers. including

All three units in beautifully finished walnut cabinets with smart aluminium vertical trims. The tuning dial is fronted by unique
black glass. Figures light up behind it in green when the unit is
on. Separate easy-to-operate Left and Right volume controls. Normal Price £115:00

on, separate easy-to-operate cert am right bothmer of the months of the controls.

Weadphone socket on front panel for easy access. Single tuning knob for FM and AM.

Built-in aerials, facilities for external FM aerial. Illuminated dial. Stereo broadcast indicator Built in parials, facilities for external FM serial, Illuminated dial. Stereo broadcast indicator
light. Each speaker has 83° bass and 3° treble units. Output 9° wasts r.m.s. per channel into
8 ohms. Frequency response: 30 to 20,000 Hz. FM: Irequency range 80-108 MHz, sensitivity 2-5µV,
stereo separation 3068, image rejection 406b.
AM: frequency range 530-1605 kHz, sensitivity 100µV.
Sizes: speakers 8. wide, 12. high and 93. deep; receiver

6" wide, 12" high and 91" deep. And their sophisticated appearance matches the excellence of the specification. 25-WATT 3-WAY CRYSLER

S.TRACK STERFO CARTRIDO TAPFORCK MODEL RP-1000ST

popular Lear-Jet type recording unit is the heart of the factoric RP 1000CT which has full record and playthe fantastic nr-100051, which has full facult and play-back facilities. Aulomatic Irack change with manual over-ride. Press to start button. Stereo headphone and Left and Right Mic sockets on front panel. Dual recording level meter, and Left and Right volume controls, Built-in premeter, and Left and Right volume controts. Built-in pre-amp. Taps speed 33 jos (95 cms.) Frequency (response: playback 30-10,000 Hz; recording/playback 30-8,000 Hz. Line output: fully variable 0-500m/. The RP-1000ST in-corporates all the features you'd expect in a top quality Cartridge Tape Deck. Size 16" wide, 4" high, 9" deep. Cabinet in walnut, Including connecting leads

£39.95

M. TRACK HOME STEREO CARTRIDGE PLAYER With this unit, you can play any standard 8-track cartridge

with (nis unit, you can pray any status of the normal retail value!

It gives you a total of 5 watts of power, to feed into two
8-ohm speakers. The frequency response is 50 to 10,000 Hz, giving you a fine tonal quality that can't be bettered at anything near this price. The E1 has separate tone, balance and volume controls, giving you complete free-dom to select the sound you want to hear. Tape speed is 3½ ips, and wow and flutter are both less than 0.3%. Size: 111 wide, 5" high, 11" deep.

ROC PRICE

TINING TILLIO. SPEAKER CE-56 This high quality speaker has its own built-in 3-way sound response switch giving you the ideal frequency response for

hi-ti natural or mond music listening lits beautiful, heavy, oiled walnut cabinet incorporates two separate speaker units an 8" woofer, and a 5 mid-renge with 2" concentric tweeter. Power handling

capacity: 25 watts r.m.s into 8 ohms. Overall frequency response: 35-20,000 Hz. Cabinet size: $10\frac{1}{2}^n \times 7\frac{1}{2}^n \times 8\frac{3}{4}^n$. Example 10 to × 7½" × 8%". Exactly right for matching the most modern decor

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PALACE AM/FM/MPX STERED TUNER AMPLIFIER SSA-16

This is one of the lowest priced stero tuner amplifiers on the market. It covers the full range of both AM and FM broadcast frequencies. And when you're switched to FM, an indicator lights up when a stereo signal is received - that's the time to switch to 'Stereo'! The SSA-16 has all the facilities you'd exswitch to Settled The SSA-10 has all the relatives you lear-pect to find on tuners costing twice as much — separate vol-ume, bass, trable, balance and tuning controls. Salector switch for tape, phono, AM. FM, stereo, Jack socket on front panel for stereo headphones. Frequency range: FM 88-108 MHz. AM 535-1605 MHz. Frequency responses: 50-10.000 Hz. ± 36B. Power output: 4 watts total music power into two 8 ohm speakers, Size: 16" wide, 41" high, 8" deep.

ROC 7-WATT STEREO AMPLIFIER CHASSIS SK-317 This exclusive Chassis is completely self-contained it costs £2-25 less than the normal re tail value! Th SK-317 is a really The compact unit measuring only 51" wide, 13"

high and 63" does it contains its own mains power supply, and has a ganged tone control and separate volume controls for each channel Specification; frequency response 40-17,000 Hz ± 3dB; output 3-5 watts music power per channel into 8 ohms; input, phono, 600mV al-to-noise ratio better than 45dB

18.25 OLSON AM-372 16-WATT STEREO AMPLIFIER 🕰

Here's a really good amplifier at a really down-to-earth price — nearly £7 Jess than the normal retail value! Just look at what the AM-372 will do for you — reproduce signals from ceramic or crystal carridges, AM and FM tuners, and tapa recorders. And it gives you outputs for two sets of hones and tape recorders. Frequency response is 30 to 20 000 Hz + 3dB. Dutput 8 watts r.m.s per channel music power into 8 ohm speakers. Phono input 200mV. Tuner input 200mV. Size: 121" wide, 31" high, 71" deep.



STEREO TUNER Here's another unit that gives

Designed specifically to match the Realistic SA-100B in both

OLSON AM-395

40-WATT STERED An ideal unit for your new

stereo separate system. It is more than £10.00 below the normal price! Making the AM-395 one of Britain's best hi-li

magnetic or ceramic pick-ups. buys. It takes in signals from magnetic or ceramic pick-ups, tuners (see Gloon RA-310) and tapa decks. And it's got outputs for taping and for headphones. There are separate bas and treble controls, separate Laft and Right channel volume controls. And a loudness switch for boosting the bass and

treble notes when listening at low output levels. Frequency response 20-20,000 Hz \pm 3dB. Dutput: 20 watts r.m.s. per channel abons. Inputs: magnetic phono 3-0mV RIAA. crystal phono 100mV; tape 150mV; tuner 150mV. Size $11\frac{1}{12}$ " wide, 4" high, $7\frac{1}{2}$ " deep. The specification reads well – sounds even better!

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R.328 STERFO HEADPHONE If you're starting in hi-fi, and you discover the need for a pair of really good stereo head-phones. The R.328 is ideal, at a price you can afford. They have

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Smart black, tough, plastic cases, containing a high flux 110mm diameter spaaker unit. Just what you need to go with the CS.8 Cartridge Player or any other car sterno system. Larrings rrayer or any other car acres on extending the fitted with over three yards of connecting cable. O'mensions: 64 " > 54" 35," impedance: 8 ohms per speaker. Rating; 5 warts max per speaker, ROC PRICE 63-72

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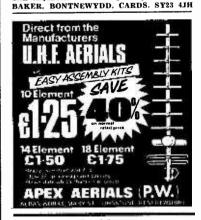
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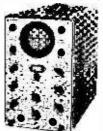
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ı	OAZ	0.88	0AQ5 0.88	OF1 0.75	7Y4 0.65	First Qu	ality E	ully Gua	rantaadi		0.65	GZ33 0.70 I	PL36 0.35)	U281 0.40
ı	OA3	0.45	6AQ6 0.65	6F5 0.75	9BW6 0.60	LIISI MA	alily r	ully Gua	IGIILEEU	EF183	0.80	GZ34 0.60	PL81 0.50	U282 0.40
1	OB2	0.85	6AR5 0.45	6F6G 0.85	10C2 0.50	-		00000			0.85	HABC800.45	PL82 0.45	U301 0.40
1	OB3	0.70	6AR6 0.55	6F11 0.40	10D1 0.55		200	79/2			1.25	HK90 0.40	PL83 0.45	U403 0.60
1	OC3	0.88	6AR11 1.25	6F13 0.45	10D2 0.50	927	ALC: U	V.			0.30	KT66 2.05	PL84 0.40	U404 0.60
1	OD3	0.85	6AS5 0.45	6F14 0.70	10F1 0.75		a . Tillis	A C			0.50	KT88 2.00	PL302 0.85	U801 0.80
ı	1B3GT	0.40	6AS7G 0.85	6F15 0.65	10F9 0.60	7.4		U BR/	MD		0.50	N78 1.50	PL504 0.80	UABC80
ı	1L4	0.20	6AT6 0.35	6F18 0.50	10F18 0.50		AND RESIDENCE OF THE PARTY OF T	DU/	MD		1.60	PABC800.40	PL508 0.90	0.40
ı	1R4	0.45	6AU6 0.25	6F23 0.85	10L1 0.50	- 1-					0.60	PC86 0.60		
ı	1R5	0.40	6AV6 0.80	6F24 0.75	10LD11 0.65		aeru							UAF41 0.50
1	184	0.80	6AW8A 0.60	6F25 1.00	10P13 0.60						0.65	PC88 0.60	PL801 0.80	UAF42 0.55
I	185	0.80	6BA6 0.85	6F26 0.85	10P14 1.10	FIEC	TRON	IC VA	LVES		0.55	PC97 0.50	PL802 0.96	UB41 0-60
1	1T4	0.30	6BE6 0.30	6F28 0.65	12AB5 0.70	ヒレヒし		IO VA	LTLS		0.42	PC900 0.48	PM84 0.60	UBC41 0.50
ŧ	1U4	0.30	6BF5 1.00	6GK6 0.60	12AC6 0.50		****** * **	1 75 77 40	T0000 A FA		0.25	PABC800-40	PY31 0.80	UBC81 0.40
1	1U5	0.60	6BF6 0.55	6J4 0.50		30A5 0.50	50EH5 0.60	DK40 0.60	ECC89 0.50		0.48	PCC84 0.40	PY33 0.68	UBF80 0.40
1	1V2	0.50	6BH6 0.75	6J5GT 0.30		30AE3 0.40	50L6GT0.55	DL96 0.45	ECC91 0.80		0.40	PCC85 0.40	PY80 0.40	UBF89 0.85
1	1X2B	0.50	6BJ6 0.50	6J6 0.20	12AE6 0.55 12AL5 0.50	30C1 0.80	85A2 0.50	DM70 0.55	ECC189 0.65		0.88	PCC88 0.55	PY81 0.80	UBL1 0.60
ł	2A3	0.50	6BK7A 0.60	6J7 0.45		30C15 0.80	90AG 2.40	DM160 0.65	ECC807 1.25		0.85	PCC89 0.55	PY82 0.85	UBL21 0.65
ı		0.70	6BN5 0.48	6K6GT 0.60	12AQ5 0.50 12AT6 0.80	30C17 0.90	90AV 2.50	DY86 0.82	ECF80 0.85		0.60	PCC189 0.55	PY83 0.88	UC92 0.40
ł	2D21	0.88	6BN6 0.45	6K7 0.85	12AT6 0.80 12AT7 0.85	30C18 0.80 30F5 0.85	90C1 0.65	DY87 0.88	ECF82 0.85	EL822		PCC805 0.85	PY88 0.40	UCC85 0.40
ł	3A4	0.40	6BQ5 0.25	6K8G 0.40	12AU6 0.35		90CV 2,40	DY802 0.85	ECF86 0.65	ELL80		PCC806 0.80	PY500 1.00	UCF80 0.55
1	3A5	0.75	6BR8 0.70	6K25 0.75	12AU7 0.80	30FL1 0.75	807 0.50	E88CC 0.65	ECF8041.65		1.00	PCF80 0.80	PY800 0.40	UCH21 0.60
1		2.00	6B87 1.80	6L6GT 0.45	12AV6 0.40	30FL12 1.20 30FL14 0.85	813 8.75	E180F 1.00	ECH35 1.00	EM71	0.80	PCF82 0.85	PY801 0.50	UCH42 0.70
ı		2.50	6BW6 0.85	6L7 0.40	12AV7 0.65	30L1 0.40	866A 0.75 5642 0.70	EABC800.85	ECH42 0.75	EM80	0.45	PCF84 0.60	PZ30 0.85	UCH81 0.40
ı	384	0.35	6BW7 0.80	6L18 0.45	12AX7 0.80	30L15 0.85		EAF42 0.55	ECH81 0.80	EM81	0.60	PCF86 0.60	QQVO2-6	UCL81 0.60
ı	5R4GY		6BX6 0.25	6LD20 0.50	12AY7 0.75	30L17 0.80	5670 0.50 6080 1.60	EBC33 0.50 EBC41 0.55	ECH83 0.45	EM83	0.50	PCF87 0.90	2.25	UCL82 0.85
ı		0.85	6BZ6 0.40	6N7GT 0.45	12B4A 0.60	30P12 0.80	6146 1.60	EBC81 0.30	ECH84 0.45 ECL80 0.45	EM84 EM85	0.85	PCF801 0.50	QQV03-10	UCL83 0.60
ı		0.45	6C4 0.88	6P1 0.60	12BA6 0.40	30P19 0.85	6146B 2.50	EBF80 0.40	ECL81 0.50	EM87	1.00 0.70	PCF802 0.50	1.25	UF9 0.60
ı	5Y3GT		6C5GT 0.40	6P28 0.65	12BA7 0.40	30PL1 0.75	6360 1.25	EBF83 0.40	ECL82 0.85	EN91		PCF805 0.80	QQV03-20A	UF11 0.50
ı	5Z3	0.60	6CB6 0.85	6Q7 0.40	12BE6 0.40	30PL13 0.98	6939 2.25	EBF89 0.80	ECL83 0.70	EY51	0.88 0.40	PCF806 0.70 PCF808 0.85	5.25	UF41 0.60
1		0.40	6CD6GA1.25	68A7 0.40	12BH7 0.45	30PL14 0.90	7199 0.80	EC53 0.50	ECL84 0.55	EY80	0.55	PCH2000.70	QQV06-40 A	UF42 0.60
1	6/30L2		6CG7 0.55	68G7 0.40	12BY7 0.60	35A3 0.55	7360 2.00	EC86 0.60	ECL85 0.55	EY81		PCL81 0.50	TT21 8.00	UF43 0.60
1		0.85	6CH6 0.60	68K7 0.40	12K5 0.70	35A5 0.75	7586 1.25	EC88 0.60	ECL86 0.40	EY83	0.40 0.55	PCL82 0.85	TT21 8.00 TT22 8.20	UF80 0.85 UF85 0.40
1	6AF4A	0.55	6CL6 0.55	68L7GT0.85	12K7GT0.40	35B5 0.65	7895 1.25	EC90 0.88	EF40 0.50	EY86	0.40	PCL83 0.65	TY2-125	UF89 0.40
١	6AG5	0.22	6CW4 0.65	68N7GT0.85	12Q7GT0.40	35C5 0.50	9002 0.40	EC92 0.85	EF41 0.65	EY87	0.48	PCL84 0.45	10.00	
ı	6AG7	0.40	6CY5 0.45	68Q7 0.40	198R7 0.40	35D5 0.75	9003 0.50	EC93 0.55	EF42 0.70	EY88	0.48	PCL85 0.40	U18/20 0.75	UL84 0.40
ı	6AH6	0.50	6CY7 070	68R7 0.40	20D1 0.50	35L6GT 0.50	AZ1 0.55	EC8010 2.25	EF80 0.25	EZ40	0.50	PCL86 0.45	U25 0.80	UM84 0,25
ı	6AJ8	0.80	6D3 0.50	6T8 0.85	20L1 1.10	35W4 0.85	AZ31 0.55	ECC35 1.00	EF83 0.55	EZ41	0.50	PCL88 0.90	U26 0.80	UY1N 0.50
ı	6AK5	0.85	6DC6 0.80	6U4GT 0.65	20P1 0.50	35Z3 0.70	CBL1 0.90	ECC40 0.65	EF85 0.85	EZ80	0.27	PCL800 0.98	U31 0.60	UY11 1.00
ı	6AK6	0.60	6DK6 0.50	6U8A 0.40	20P4 1.10	35Z4G 0.85	CBL31 1.00	ECC81 0.85	EF86 0.80	EZ81	0.29	PCL801 0.75	U37 2.10	UY41 0.48
1	6AL3	0.48	6DQ6B 0.70	6V6GT 0.40	20P5 1.20	35Z5GT 0.60	CY31 0.35	ECC82 0.80	EF89 0.28	GTIC	8.00	PD500 1.80	U52 0.85	UY82 0.50
١	6AL5	0.20	6EA8 0.60	6X4 0.85	25C5 0.50	50A5 0.75	DAF96 0.45	ECC83 0.80	EF91 0.88	GY501	0.80	PF86 0.65	U76 0.85	
1	6AM5	0.85	6EH7 0.80	6X5GT 0.40	25L6GT 0.50	50B5 0.50	DF96 0.45	ECC84 0.80	EF92 0.85	GZ30	0.40	PF818 0.85	U78 0.85	
1			6EJ7 0.85 6EW6 0.70	6X8 0.60 6Y6G 0.70	25Z4G 0.80	50C5 0.50	DK92 0.65	ECC85 0.40	EF95 0.85	GZ31	0.85	PFL200 0.65	U191 0.75	W729 0.75
1	6AM6	0.88	6EW6 0.70	6Y6G 0.70	25Z6GT 0.65	50CD6G1.20	DK99 0.60	ECC88 0.40	EF97 0.65	GZ32	0.48	PL33 0.85	U201 0.35	Z803U 1.20

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2N3416	0.25	BC147	0.175	OC78	0.25		
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2N3702 2N3703	0.12	BC152	0.15	OC81D	0.25		
2N3704	0.17	BC158	0.15	OC83	0.20		
2N3707	0.15	BC175	0.20	OC139	0.80		
2N3709	0.12	BC186	0.25	OC140	0.85		
2N3710 2N3819	0.1 <u>2</u> 0.35	BCY30 BCY31	0.25 0.40	OC141 OC170	0.60 0.25		
2N3906	0.20	BCY33	0.25	OC171	0.25		
28702	0.50	BCY34	8.80	OC200	0.30		
28746	0.25	BCY72	0.20	OC201	0.60		
AC113 AC125	0.15 0.80	BCZ10 BCZ11	0.80	OC202 OC203	0.65 0.40		
AC126	0.20	BD121	0.40 0.65	OC203	0.40		
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3-in tube fitted with telescopic viewing hood, giving bright display in full daylight. Sensitivity 10mV/mm (narrow band) to 30mV/mm (wide band). Bandwidth 10 c/s--10 me/s. Triggered aweep pre-set at 1-2-5-10-30-100-300-100-3000 µsec per stroke. Free-running time base 20 c/s to 20000kc/s with built-in crystal calibrator providing timing marks at .05-,2-1-5-20-100 µ sec.
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Input attenuator 1-10-100. Power supplies 127/230v AC.

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INTEGRATED CIRCUITS TAA263 Direct coupled 3-stage amplifier up to 600kc. Supply voltage 6-8V output 10mW	
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sipation 200mW. TO18 outline	0.65
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MULTIMETERS

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TYPE U4312

Low sensitivity (667 ohms per volt), 1% accuracy DC and 1.5% AC. 9 DC current ranges 300µA to 6A, 8 AC current ranges (1.5mA to 6A). 9 AC/DC voltage ranges 0-3-900V. Resistance 0-2-30K.

TYPE U4313. High sensitivity 20 Kohms/voit DC and 2 Kohms/voit AC. Accuracy 1-5% DC and 2-5% AC S DC current ranges 60,4th to 1-5A. 6 AC current ranges 0-60mA to 1-5A. 9 AC and DC Voltage ranges 1-5 to 50V. Resistance ranges 0-5-500 Kohms. 410-50

TYPE U4341 MULTIMETER AND TRANSISTOR TESTER



With taut hand suspension move-

ment.

8-DC voltage ranges 0·3-900V

6-AC voltage ranges 1·5-750V

5-DC current ranges 0·6-600mA

4-AC current ranges 0·3-300mA

4-resistance ranges 0·6-600 Kohms

Transistor cut-off current 60µA max

Transistor DC current gain 10-350

Sensitivity: 16700 QV DC;

3300 Q/V AC PRICE 210-50

TYPE U435



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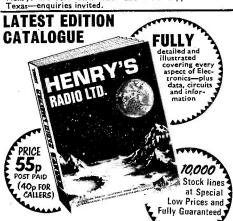
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