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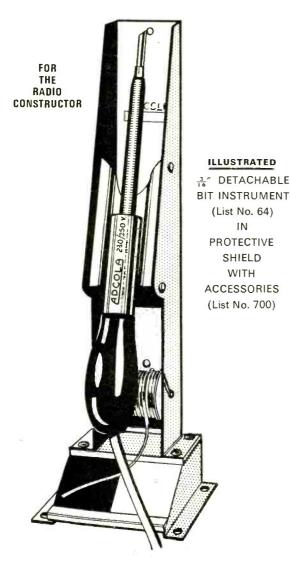


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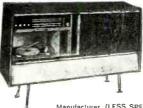
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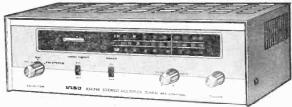
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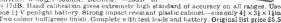
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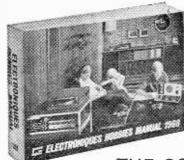
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AC126 4/		A 5/-	OC75	4/6	XB112	3/-
AC127 4/	6   GET10	6 5/6	OC76	4/6	XC141	7/-
AC128 4/	6   GET11	3 5/6	OC77	4/6		
ACY19 5/	- GET87	3 5/6	OC78	5/-	PHOTO	-
ACY21 5/	- GET87	4 4/6	OC81	. 4/-	TRANS	18-
AD140 8/	6 MAT10	0 7/9	OC81M	2/6	TORS	12/6
AD149 16/-	- MAT10	1 8/6	OC81D	4/-		
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15/-			OC82	4/6	SIGN	AL
AF114 6/		03	OC83	5/-	DIOD	E.C.
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AF116 4/	6 OC23	8/6	OC140	11/-	IN34A	4/-
AF117 4/	6 OC26	7/6	OC169	4/6	OA5	3/6
AF118 4/	6 OC28	8/6	OC170	4/6	OA70	2/-
AF119 4/	B OC35	8/6	OC171	4/6	OA79	2/-
ASY28 7/-	- OC41	5/-	OC200	9/-	OA81	2/-
BC107 5/	B OC42	5/-	OC201	11/-	OA90	2/6
BC108 5/	B   OC44	4/-	OC203	12/6	OA95	2/6

#### **ZENER DIODES**

OAZ20012/- OAZ20110/- OAZ202 8/6 OAZ203 8/6	OAZ204 OAZ205 OAZ206 OAZ207	8/6 8/6 8/6 9/6	OAZ208 OAZ209 OAZ210 OAZ211	6/6 6/6 6/6 6/6	OAZ212 OAZ213 OAZ2271	6/0 6/0 5/-

STC. 1 WATT SERIES 5% 2.4/2.7/3/3.9/4.3/18/16/18/20/30/33 volt, 5/\* each. Z series. All voltages from 3.9-50 volt. 250 mW, 2/8 ea. 1.5 w.4/\* ea. 7 w.5/\* each.

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CAPACITY ANALYSER 2 pF-2,000 mFd 2 ohms 200 megohms. Also checks impedance, turns ratio, insulation, 200/250V A.C. a.C. prand New £17.10.0. Carr. 7/6.



#### T.F.40 HIGH SENSITIVITY A.C. VOLTMETER

10 meg. input 10 ranges: -01/-003/-1/-31/3/10 / 30 / 100/300V. R. M. B. 4cps. -1-2Mc/s. Decibels -40 to +50 dB. Supplied brand new complete with leads and instructions. Operation 230V. A.C. \$17.10.0. Carr. 5/-.



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Brand new with knobs as 0/115/230V. Step up or step down. Fully

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AF. SINE W 20-200,000 Square wave 30,000 c/s. HIGH IMP. WAVE 30,000 c/s. O/P HIGH IMP. 21V P/P600Ω 3·8V P/P. TF 100Kc/s-300 Mc/s. Variable R.F.

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**POWER DIODES** 

290 MA 2/2 70 P.I.V. 1 AMP 3/6 140 P.I.V. 165 MA 1/-165 P.I.V. 200 P.I.V. 200 P.I.V. 200 P.I.V. 500 MM 5/6 800 P.I.V. 500 MM 5/6 800 P.I.V. 6 AMP 7/6 400 P.I.V. 6 AMP 7/6 100 P.I.V.

**THYRISTORS** SILICON CONTROL

RECTIFIERS

400 P.I.V. 3 AMP 7/6 100 P.I.V. 7 AMP 13/6 200 P.I.V. 7 AMP 15/6 400 P.I.V. 7 AMP 15/6

**PLEASE** 

ADD

6 AMP 5/6 400 P.I.V. 8 AMP 7/6 700 P.I.V.

650MA 6/6

60 P.I.V. 290 MA

400 P.I.V.

attenuation int/ext. modulation. Incorpor ates dual purpose meter to monitor AF out-put and % mod. on R.F. 220/240V A.C. £30.0.0. Carr. 7/6

#### TE-20RF SIGNAL GENERATOR



for details.

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Just released by the Ministry, High quality 10 valve receiver manufactured by Murphy. Coverage in 5 bands 650 Kc/s-30 Mc/s. 1.F. 500 Kc/s. Incorporates 2 R.F., and 3 I.F. stages, bandpass filter, noise limiter, crystal controlled B.F.O., calibrator. I/F. output, etc. Built-in speaker, output for phones, Operation 150/230 voit A.C. Size 19½ x 134 x 16in. Weight 114 lb. Offered in good working condition. \$22.10.0. Cerr. 30/-. With circuit diagrams. Also available B.41 L.F. version of above. 15 Kc/s-700 Kc/s. \$21.71.00. Cerr. 30/-.

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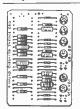
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+ 22dB. 10.0, 100mFd. 0.100-0.1mFd. 89/8 P & P 2/6

250J.

0/10/50/500/ D.C. 0/10/50/

v. A.C. 0/250

MODEL

500/2,500

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Equally adaptable for checking diodes, etc. Spec; A 0.7-ing diodes, etc. Spec; A 0.7-ing diodes, etc. Spec; A 0.7-o.9967. B: 5-200, lec 0.9-50 microamps 0-5mA. Resistance for diode 200 Q -1M Q. Supplied complete with instructions, battery and leads, £5.19.6. P. & P. 2/6.



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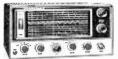
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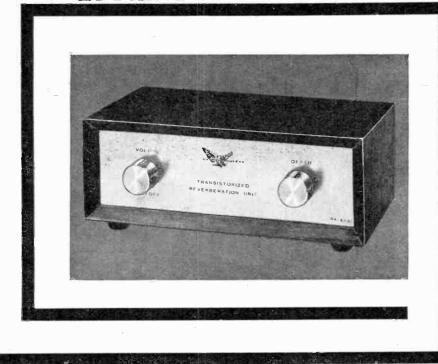
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Model A1005M, Simple instructions to convert any FM tuner. Max input 100 m volts. Power supply 8-14 volts. Mains or batt. (5mA at 12V.) Output 1.5 x input. 4 transistors (2 each) 2SB-2O2: 2SB-186. 7 idiodes IN34 (six) IN60 SAVE 34/-5 Gns.

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Sin.	150ft.	2/3	3in.	225	ift.	2/9
4in.	300ft.	4/6	4in.	450	ft.	5/6
5in.	600ft.	7/6	5in.	900	ft.	10/6
5fin.	900ft.	10/6	5%in.	1,200	ift.	13/
7in.	1,200ft.	12/6	7in.	1,800	ft.	18/6
	Double Play			Tripl	e Play	
8in.	300ft.	4/-	4in.	900	ft.	13/-
4in.	600ft.	8/-	5in.	1,800	ft.	25/-
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#### STEREO PORTABLE CABINETS

Latest black and silver metal finish. Consisting of centre cabinet size 16-tim. x 13 in. x 8 in. deep with lift up lid together with two 10 x 6 speaker cabinets which clip on ends of main cabinet size 4 in. x 13 in. x 8 in. making overall size of 25 i in. x 13 in. x 8 in. making overall size of 25 i in. x 13 in. x 8 in. high quality chrome 6ttings. Will take almost any autochanger or tape deck.

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PLEASE NOTE. A wide range of cabinets
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100 HI-STABS 9/-1% to 5%  $100\,\Omega$  to  $5~{\rm m}\,\Omega$ . CO-AX, low loss, 6d. yd., 25 yds. 11/6; 50 yds. 22/-; 100 yds. 42/6. Plugs 1/3.

100 RESISTORS 6/6 SIZES—i-3 watt, MICROPHONE CABLE. Highest quality, black, grey, white, 9d. per yard.

100 CONDENSERS 9/6 Miniature Ceramic, Silver, Mica etc., 3pF to 5µF. LIST VALUE OVER £4,

25 ELECTROLYTICS Assorted 2 to 500 mfd 6 to 400 volt. LIST VALUE OVER £5. 9/6 **50 TAG STRIPS** 

Mixed sizes 2 to 15 way. **25 POTENTIOMETERS** Including with switch, long and short spindle, pre-sets, log and lin. unused,  $1k-2m\Omega$ . VALUE APPROX. 24. 9/6

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#### **VIKING AMPLIFIER**

#### 50 WATT AMPLIFIER

An extremely reliable general purpose valve amplifier. Its rug-ged construction yet space age styling and design makes it by far the best value for money

TECHNICAL SPECIFICATIONS

TECHNICAL SPECIFICATIONS

4 electronically mixed channels, with 2 inputs per channel, enables the use of 8 separate instruments at the same time. The volume controls for each channel are located directly above the corresponding input sockets.

SENSITIVITIES AND INPUT IMPEDANCES

Channel 1 4mV at 470 K

Channel 2 4mV at 470 K

Channel 3 200mV at 1m

Channel 3 200mV at 1m

Channel 4 200mV at 1m

Channel 4 200mV at 1m

Channel 5 4mV at 60 Hz/s.

Suitable for most high output instruments

(gram, tuner, organ etc.).

Input sensitivity relative to 10w output.

TONE CONTROLS ARE COMMON TO ALL INPUTS

Bass Boost + 12dB at 60 Hz/s.

Treble Boost + 11dB at 15 KHz/s.

Treble Cut - 12dB at 15 KHz/s.

With bass and treble controls central - 3dB points are 30 Hz/s and 20 KHz/s.

POWER OUTPUT

For speech and music 50 watts rms. 100 watts peak. For sustained music 45 watts rms. 90 watts peak. For sine wave 38.5 watts rms. Nearly 80 watts peak.

Total distortion at rated output 3.2% at 1KHz/s
Total distortion at 20 watts 0.15% at 1KHz/s
Output to match into 8 or 15 ohms speaker system.

P & P 20/
MEGATIVE FEED BACK 20dB at 1KHz/s. SIGNAL TO NOISE RATIO
60dB. MAINS VOLTAGES Adjustable from 200-250v A.C. 50-60 Hz/s.

A protective fuse is located at the rear of unit.

VALVE LINE UP Double purpose ECCs3 x 3, EL34 x 2 and GZ34.

#### FOUR PLUS FOUR Stereo Amplifier

the great demand we are now able to offer this precision made instrument at a fantastically low price. The high quality, reliability and styling has been maintained in spite of its low price.

SPECIFICATIONS

16"

SPECIFICATIONS
Elegant styled cabinet (sizes 16" wide 5" high 8½" deep) in black rexine and woodgrained sides. Brushed aluminium front panel with contrasting black/silver

CONTROLS Stereo/Mono switch. Gram/Aux switch. Volume left. Volume right. Treble (cut and lift). Bass (cut and lift). Separate on/off switch. Neon pilot indicator.

INPUTS AND OUTPUTS

(per channel) Gram, aux, tape out and speaker out. A switched mains socket is

A superb high quality, yet inexpensive stereo amplifier. Due to the great demand we are now able to offer this precision made instrubridge rectification.

Price

27 gns

Price

13 gns

P& P 15/-

TECHNICAL SPECIFICATIONS
Gram sensitivity 40mV at 1 KHz.
Aux sensitivity 50mV at 1 KHz. (Sensitivities are given for rated output). 4 watts r.m.s.

(8 watts r.m.s. in monaural position). Output matches in-to standard 3 ohins speaker system. Suitable 10" x 6"

speaker system. Suitable 10" x 6" peakers are available at 29/6 each, plus 5/- p. & p. Bass control at 100 Hz lift + 9dB cut — 10dB. Troble control at 10 KHz lift + 8dB cut — 13dB. Total harmonic distortion 0.35% at 3 watts and 2% for rated output at 1 KHz. Negative feedback 13dB at 1 KHz. Mains supply 220-250V A.C. 50-60 Hz.

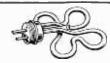
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Size 3\frac{1}{2} x 1\frac{3}{4} in Meter size 2\frac{1}{2} x 1\frac{3}{4} in Sensitivity 1000 O.P.V. on both A.C. and D.C. volts. 0-15, 0-150, 0-1000 D.C. current 0-150mA, Resistance 0-100k\Omega. Complete with test prods, battery and full instructions, 42(6, P. & P. 3/6, FREE GIFT for limited period only. 30 watt Electric Soldering Iron value 15/- to every purchaser of the Pocket Multi-Meter.

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Complete with PC88 and PC86 Valves. Full variable tuning. New and unused. Size 42×51× 14in. Complete with circuit dlagram. 35/- p. & p. 3/6



#### THREE-IN-ONE HI-FI 10 WATT SPEAKER

10 WATT SPEAKER

A complete Loud Speaker system on one frame, combining three matched ceramic magnet speakers with a low loss cross over network. Peak handling power 10 watts. Impedance 15 ohms. Flux density 11,000 gauss. Resonance 40-60 c/s. Frequency range 50c/s to 20kc/s. Size 13\footnote{1}\text{x} 2\footnote{1}\text{hree} x y famous manufacturer. List Price £7. Our price 69/6 plus 3/6 P. & P.

Similar speaker to the above without tweeters in 3 and 15 ohm 39/6 plus 3/6 P. & P.

#### 600 mW FOUR TRANSISTOR AMPLIFIER

Features N.P.N. and P.N.P. complementary symmetrical output stage,  $24^{\circ} \times 4^{\circ} \times 4^{\circ}$  Speaker. Output impedance 12 ohms frequency response 3dB points 90 c/s and 12 Kc/s. Price 19/6 plus 1/- P. & P. 7 x 4" Speaker to suit, 13/6 plus 2/- P. & P.

#### 2½ watt ALL TRANSISTOR AMPLIFIER

AC mains 240V. Size 7" x 4\frac{1}{2}". Frequency response 100 c/s—10 Kc/s Semi conductors, two OC 75's two AC 128's and two stabilizers AA129. Tone and volume controls on flying leads. £2.10.0 plus P. & P. 3/6. Suitable 8" x 5" 10,000 line high flux speaker, 18/6 plus 2/- P. & P.

**NEW TRANSISTORISED** SIGNAL GENERATOR

Size  $54 \times 34 \times 14$  in. For 1F and RF alignment and AF output 700 c/s frequency coverage 460 kc/s to 2 mc/s in switched frequencies. Ideal for alignment to our Elegant Seven and Musette. Built and tested. 49/6 P. & P. 3/6. & P. 3/6.

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#### THE RELIANT 10W SOLID-STATE HIGH QUALITY AMPLIFIER

SPECIFICATIONS

Output Impedance-3 to 4 ohms

Output—10 watts RMS Sine-wave 13 watts RMS Music-power Inputs—1.-xtal mic 10mV 2.-gram/radio 250mV

Tone Controls—Treble control range  $\pm~12 dB$  at 10 KH2Bass control range  $\pm~13 dB$  at 100 H2

2. •gram/radio 250mV Bass control range  $\pm$  13dB at 100H2 Frequency Response—(with tone control central) Minus 3dB points are 20H2 and 40KH2. Signal to Noise Ratio—better than -60dB. Transistors—4 silicon Planar type and 3 Germanium type. Mains input—-220—250V. A.C. Size of chassis— $10^7 \times 34^7 \times 2^7$ . A.C. Mains, 200-250V. For use with 8d or L.P. records, musical instruments, all makes of pick-ups and mikes. Separate bass and treble lift control. Two inputs with control for gramand mike. Suit and tested.  $8^7 \times 5^7$  speaker to suit price 14/6 plus 1/6 P. & P. Crystal mike to suit 1/6 P. & P. & P. & P.

#### LESLIE TREMOLO SPEAKER UNIT

Model 10C3. Made in U.S.A. by Electro Music Corp., Pasadena, Calif. Complete with 8' 6 ohm Heavy Duty Loud Speaker. (Larger Speaker may be fitted without major structural change). Dimensions: baffle size 18\frac{1}{2}' \times 19'. Rotor size 15'. A.C. mains 110-220-240 50 cfs. suitable for organs.

PRICE £30 plus £2 P. & P

PRICE £30 plus £2 P. & P

#### 50 WATT POWER AMPLIFIER

50 WALL PUVVER AIVIFLIFIED

Manufactured by Livingstone Labs, at a cost price of £90. In two units, Amplifier size 10½" × 9" × 7". Heat sink size 14½" × 9" × 4". Output into 15 obm Speaker. Transistors—eight OC28s. Two 26374. Sensitivity 1V. RMS at 6K input impedance. Power Response 5H2-15KHZ (50 watts). Ideal for Tannoy Systems or organs where constant use is essential.

PRICE 28gns. plus £2 P. & P.

#### ANTI-THIEF CAR BURGLAR ALARM

The Meiguard Safermatic consists of an electrical device housed in small metal box  $4^r \times 2^r \times 1^{p^r}$ , which has been designed and developed to provide protection required by the average motorist at an economic cost. Using this system, an alarm and the immobilised condition is set automatically as soon as you park the car. Should you leave the key in the ignition, no one but you can drive the car away. Upon entering the vehicle the method of starting the car is by switching on the ignition, depressing two hidden switches, and simultaneously operating the starter. Location of the switches is known only to you. Should the alarm be set off it can be stopped by following the normal starting procedure. For 12v operation, List price 79/6, our price 29/6 plus 2/6 P. & P. Full easy-to-follow instructions supplied.

#### EXTRACTOR FAN



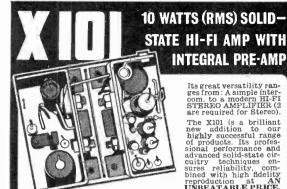
AC Mains 230/250 v. complete with pull switch. Size 6" x x Price 27/6 plus 5/-P. & P.



**AC MAINS** MOTOR 1400 R.P.M. 230/250v

> PRICE 9/6

P. & P. 3/-



SPECIFICATIONS R.M.S. Power Output: 13 W (music power), 10 W (Sine Wave). Sensitivity: for rated output 1 mV into 3 K ohms load. Frequency Response: minus 3dB points are 20 Hz and 40 KHz. Total Distortion: at 1KHz for rated output 1.5%, for 5 W output

Output Impedance: 3 ohms (3-15 ohms may be used).

Supply Voltage: 24 V D.C. at 800 mA (6-24 V may be used) output at 14 V D C. supply with 3 ohms speaker. 7 watts.

Size: 28' x 3' x 1\*/16.

The fully comprehensive instruction manual does not only show the basics, such as circuit diagram and connections, but also gives practical easy-to-understand detailed information about the X101. Standard equalisation networks are given for most types of conventional inputs. They include: Tape Head, Mag. P.U., Xtal P.U., Tuner, Mic, etc. 49/6 + 2/6 p. & p.

Control assembly: (Including resistors and capacitors).

1. Volume: PRICE 5/-, 2. Treble: PRICE 5/-,
3. Comprehensive bass and treble: PRICE 10/-,
The above 3 items can be purchased for use with the XI01.

#### POWER SUPPLIES FOR THE X101

P101 M (formono) 35/-

P101 S (for stereo) 42/6

reproduction at AN UNBEATABLE PRICE.

#### A HIGH QUALITY MONAURAL PRE-AMP & CONTROL UNIT

Particularly suitable for use with the X101 if a ready-built, comprehensive, multi-input system is desired.

#### CONTROLS

Selector Switch, Tape Speed Equalisation Switch (3% and 7% i.p.s.), Volume, Treble, Bass, 3 position scratch filter and 3 position rumble filter.

#### SPECIFICATION

Sensitivities for 200 mV output at 1 KHz.

Sensitivities for 200 mv output at 1 kHz.

Tape Head: 3 mV (at 3\(\) i.p.s.)

Radio; 100 mV

Mag. P.U.: 2 mv
Cer. P.U.: 80 mV

Tape/Rec. Output: 100 mV

Tape/Rec. Output: 100 mV

Tape/Rec. Output: 100 mV

Equalisation for each input is correct to within ±2dB (R.I.A.A.)

from 20 Hz to 20 KHz.
Treble ±1ddB at 60 Hz

Total Distortion: (for 2000 mV output) < 0 02%

Signal Noise: > -8000 mV output) < 0 02%

Supply Voltage: 24 V D.C.

70/6 plus 2/6 p. & p.

79/6 plus 2/6 p. & p.

A STEREO VERSION (PR101/S) WILL BE ANNOUNCED SHORTLY

### The CLASSIC HIGH QUALITY SOLID-STATE

#### SPECIFICATION

Switched inputs for: Tape Head, Mag. P.U., Cer. P.U., Radio and Aux.

Mains Input: 220-250V A.C. 50 Hz.

THE CLASSIC IS THE COMBINATION OF THE ABOVE DESCRIBED ITEMS (XIOI, P101/M AND PR101/M) ON ONE COMMON CHASSIS: ITS PERFORMANCE AND SPACE-AGE STYLING MAKE IT THE IDEAL CHOICE THE VALUE-CONSCIOUS HI-FI ENTHUSIAST. Size 12½in. long, 4½in. deep, 2¾in. high. Teak finished case.

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600 milliwatt solid state 7 transistor plus diode and thermistor, Completely modulised high quality portable radio featuring complementary NPN and PNP output stage. The comprehensive easy-to-follow drawings supplied make this the easiest-ever transistor radio set of paris, with the following features.

Simple connections to only 6 tags on the R.F./I.F. module, 3 L.F. stages, osc. coil and 3 transistors which with their associated components are completely wired. Only 4 connections on the A.F. module to complete the 4 transistor 500 milliwatt solid state amplifier. Pre-aligned R.F./I.F. module built and tested.
A.F. module built and tested.
Fully tumable over M.W. and L.W. bands. M.W. 540-1640 Kc/s (557-183 metres). L.W. 150-275 Kc/s (2000-1100 metres).

(2000–1100 metres).

Intermediate Frequency 470 Ke/s.

Sensitivity: M.W. at 1 Me/s 10 microvolts plus or minus 3dB. L.W. at 200 Ke/s 40 microvolt plus or minus 4dB.

High Q internal ferrite roil aerial on both wavebands.

Class B' modullsed output stage with thermistor controlled heat stabilisation. Class B' output stage ensures long battery life. Current drain is proportional to the output level. Total current drain of the receiver under no signal conditions is 10-12 mA. At reasonable listening level 20-30 mA.

Extension sockets for car aerial input, tape recorder output (Independent of volume control) and External Speaker.

All components (except speaker) mount on the printed circuit board. Easy-to-follow instructions. Size of cabinet 12in. long 8ln. high and 3in. deep.

Special Offer—Power Supply Kit to purchasers of Dorset Portable Radio parts incorporating mains transformer, rectifier and smoothing condenser, AC mains 200/250V output 9V 100mA, 9/6 extra

#### ELEGANT SEVEN Mk III

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and CAR RADIO

plus 7/6 postage and packing. Circuit Circuit and parts list 216

Buy yourself an easy to build 7 transistor radio and save at least £10.0.0. Now you can build this superb transistor superhet radio for under £4.10.0. No one else can offersuch a fantastic radio with so many de luxe star features.

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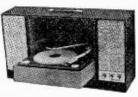
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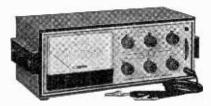
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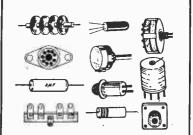
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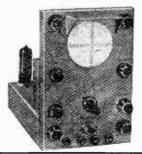
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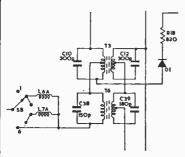


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## WIRELESS

**VOL 44** No 5

issue 739

SEPTEMBER 1968

#### TOPIC OF THE MONTH

#### Politics in P.W.

AFTER reading a certain leader, a reader fired off a broadside headed "Keep Politics out of P.W." Another, signing himself "Member of the Labour Party", obliquely suggested that we are in cohorts with the Tory Party. These, as similar missives that arrive from time to time, are just plain daft.

Politics with a capital P do not interest us. But we are concerned with the things that politicians do, or do not do, that affect radio. Without being pompous, it is our duty to speak out against acts or non-action detrimental to amateur radio, whether due to organisations, individuals, or politicians, whether they are Conservative, Labour, Liberal, Communist or East Finchley Nationalists!

We are not always right but we reserve the privilege of saying what we think. To deny this freedom of expression because it might upset political diehards is manifestly absurd. The "solidarity" we sometimes hear about really means blind faith, but we prefer to face facts.

We argued that pirate radio stations were anarchic; they have now been scuttled. We campaigned for an end to the ludicrous walkie-talkie situation; legislation now exists. We poked fun at the GPO for their gaff about morse speeds; action has been taken. Years ago we prodded the Government about local radio stations; they have now started up. We were almost the only periodical to spot the inherent dangers of Clause 7 of the Wireless Telegraphy Act; this situation was subsequently remedied. We even spoke out on mail order trading; while everything is not ideal, things have improved and due to the controversy we heard many constructive viewpoints from both customers and suppliers.

We will therefore *keep* politics in P.W., if this definition means pointing the editorial finger impartially and objectively at anything detrimental to the hobby. In the process we will, no doubt, upset political fanatics, ornithologists, taxidermists, gasfitters, the Society for This, the Committee for That and the League for The Other. Only one thing worries us—the letters of protest we are sure to get from the East Finchley Nationalists!

W. N. STEVENS—Editor.

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OCTOBER ISSUE WILL BE PUBLISHED ON SEPTEMBER 6th

All correspondence Intended for the Editor should be addressed to: The Editor, "Practical Wireless", George Newnes Ltd., Tower House, Southampton Street, London, W.C.2. Phone: TEMple Bar 4383. Telegrams: Newnes Rand London. Subscription rates, including postage: 38s. per year to any part of the world. © George Newnes Ltd., 1968. Copyright in all drawings, photographs and articles published in "Practical Wireless" is specifically reserved throughout the countries signatory to the Berne Convention and the U.S.A. Reproductions or imitations of any of these are therefore expressly forbidden.

## news and comment...

#### WALKIE-TALKIES

Unaware of the ban on manufacture and import of certain radio transmitters, a number of British travellers have bought walkie-talkies abroad and, on return to this country, have had the sets detained by Customs. The ban has applied from 1 April, 1968, and no exceptions are allowed which do not have Post Office authority.

Any use of radio in this country must have a licence and the Post Office will not licence walkie-talkies using frequencies around 27Mc/s because they are liable to interfere with authorised radio services. The Post Office are quite adamant that it will not authorise manufacture or import of transmitters which cannot be legally used here. Apart from the ban, the maximum penalties for illegally using a radio transmitter were increased last year.

Make sure before you buy a set that you can get a licence for it. If in any doubt contact the Post Office Radio and Broadcasting Dept., Armour House, St. Martin's-le-Grand, London, E.C.1.

#### **RADIO RALLIES**





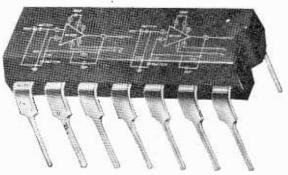
Gilwell Park in Essex was the venue of the R.S.G.B. National Mobile Rally on 23rd June. The top photograph shows the 2-metre talk-in station situated in the corner of a barn.

The lower photograph shows the topband talk-in station, GB3USA, at the A.R.M.S. Rally at R.A.F. Mildenhall held on 30th June.

This superb rally was attended by thousands, many of whom were mobile stations. Attractions included a huge trade show, "train" rides for the children, and an American Air Force band.

Highlight of the rally was a brilliant display by the Royal Air Force Aerobatic Team—The Red Arrows. Well done chaps, you were just great.

#### INTEGRATED STEREO PREAMP



A dual preamplifier, designed for high fidelity amplification of low-level stereo signals, has been announced by Motorola Semiconductors Ltd. Called the MC1303P, the monolithic device features a short-circuit-proof design, and the output leads can be shorted without harming the device. Channel separation is claimed as 60dB min. at 10kc/s with less than 0·1% total harmonic distortion. Minimum open loop voltage gain is 8,000.

#### I.R.E.C.E.

Don't bother to look it up, it stands for the International Radio Engineering and Communications Exhibition (the R.S.G.B. "do") which will be held this year at the Royal Horticultural New Hall, Greycoat Street, Westminster, London, S.W.1. Dates for your diary are 2 to 5 October.

The Diplomatic Wireless Service will be exhibiting for the first time, and another new trend will be a series of four lectures on "Communications Equipment", "Single Side Band", "Aerials" and "Communication Research and Developments". These will take place on the Thursday and Friday, with Saturday being reserved for all Armed Forces general meetings. A manufacturers' Silver Plaque will again be awarded, and some lucky visitor will win a brand new communications receiver. There will also be awards for the best home-constructed equipment.

The Exhibition is to be opened by the Postmaster-General. This might be the time for telling us just what is happening about this "Beginner's" Licence?

The three "Practicals"—Wireless, Television and Electronics—will again be taking a stand and members of the staff look forward to meeting readers.

#### S.S.S.S.

Standing for Super Stable Signal Source, and referring in this instance to the BBC's 200kc/s transmitter at Droitwich. This has long been used as a standard frequency reference by amateurs, professionals and industry. Since 1965 the long-term stability has been within  $\pm 5$  parts in 10½ and with the use of automatic frequency correction has been kept within 1 part in 10½.

The frequency control source is now a rubidium gas cell standard provided by the National Physical Laboratory. This has a day-to-day stability of better than 1 part in  $10^{11}$  and the frequency will be maintained within  $\pm 2$  parts in  $10^{11}$  of nominal referred to the caesium beam frequency standard at the National Physical Laboratory.

## news and comment...

#### **DIAL-A-COMPUTER**

An advanced data communications system which enables computers to communicate internationally over normal switched telephone lines, and makes it possible for data to be fed into a computer in another country, or even another continent, was shown at the British Engineering Exhibition in Copenhagen recently by the Marconi Company.

The system, known as Marconidata, enables data to be transferred directly into a computer at any point which can be reached by telephone. Errors due to noise and interference are virtually eliminated by the system, which will detect and correct all but about 1 in 10,000,000, equivalent to transmitting the entire works of Shakespeare twice, with only one letter in error. Now you know why examiners are against people creeping out and making telephone calls during exams!

#### ALL AT SEA





Sailing long distances single-handed seems to be catching on fast these days. The most recent activity in this field was the transatlantic race. Strangely enough the radio gear used, particularly receivers, is not unduly specialised and most of the sets used are readily available to the public.

The two photographs show the receivers used on two of the boats. The top one shows South African Brüce Dalling at the "wheel" of his Philips Transworld f.m./a.m. portable. This covers long, medium, v.h.f., plus four shortwave bands. Bruce was using a fifty-foot stainless steel vertical aerial to keep in touch with civilisation.

The lower photograph is of Eric Willis here seen massaging the controls of his receiver—a Hallicrafters CR50 covering 185–400kc/s; 535–1605kc/s; 2–5·2Mc/s; 5·2–15Mc/s and 108–135Mc/s. It also boasts separate meters for battery monitoring and signal strength. Messrs. Electroniques stock the CR50.

#### 64 DOLLAR QUESTION

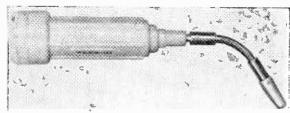


Only score 2 if you thought it was a maggot's kneecap under a microscope. Score full marks if you said it was a Clinton Prodlite. The makers inform that it will fit most 2-cell  $(2\frac{1}{2}-3 \text{ volt})$  torches and although only one-eighth of an inch in diameter, it encloses a high efficiency long lifelens bulb. The Prodlite provides illumination at otherwise inaccessible spots. It is available in two standard lengths, Model A,  $5\frac{1}{2}$  in. long, and Model B, 9in. An insulating sleeve for optional use is supplied free with each Prodlite.

A special switch handle is also available enabling the Prodlite to be used from an external power source.

Further information may be obtained from The Clinton Laboratory, 43 Broomhall Place, Sheffield, S10 2DG.

#### F.T. SPANNER



We weren't quite sure either but apparently it's a flexible torque spanner marketed by Ward Brooke & Co. Ltd., Avionics Division, Loudwater, near High Wycombe, Bucks. The units employ a unique mechanism which combines a ball bearing flexible driver shaft with a ratchet system, constructed of nylon to provide constant friction characteristics and obviate the risk of slipping. It is housed in a fluted nylon handle which helps ensure that the tool is tough enough to withstand all forms of handling. A chart is available giving the recommended loadings when torque spanners are employed on stud or screw materials having an ultimate tensile strength of not less than 40 tons/in.².

The overall length of the standard F.T. Spanner is about  $6\frac{1}{2}$  inches and is available in socket sizes from 10BA to 4BA and 4-40 to 8-32UN.

#### **NEW TRIAC**

RCA Great Britain Ltd. has extended its thyristor and triac range by introducing a new low-cost 8A triac with a moulded silicone encapsulation, the TA7365. This is a gate-controlled full-wave device with three horizontal leads to facilitate mounting on PC boards. Full-wave control of a.c. loads as in motor speeds, light dimmers, heater controls and power switches are some of its applications.





HIS set of equipment consists of a high-gain 4-transistor amplifier, with battery and loud-speaker, into which can be plugged various accessory units. A very versatile set of equipment is thus easily built up, according to the user's needs.

The amplifier can furnish up to 1-watt output, which is easily sufficient for such purposes. Accessory units include the following:

units include the following.

DIODE TUNER. This is extremely simple, but

allows excellent reception of local stations.

MORSE OSCILLATOR. This unit is for code practice; or for monitoring outgoing c.w. transmissions by rectifying r.f. to provide its operating current.

SUPERHET TUNER. Using a ferrite rod aerial, this employs three transistors and a diode in a medium wave circuit, and provides a large selection of transmissions.

RECORD DECK. This is battery operated, 33/45 r.p.m., to allow playing records with the amplifier.

TRANSISTOR T.R.F. TUNER. This is a straightforward unit with 2-gang tuning, for use when a superhet tuner is not required, and when a simple diode tuner does not give adequate results.

S.W./M.W. SUPERHET TÜNER covering 515kc/s—15Mc/s (580—20 metres) in three switched bands.

There is, of course, no need that all the accessory units be constructed. A tuner only might be required, with possibly facilities for record playing later. Or the most important interest might be in code practice, the amplifier also serving for radio reproduction later.

#### **Amplifier Circuitry**

This is shown in Fig. 1, and has enough gain for a crystal pick-up and the other units. D.C. operating conditions of each stage are independent of other

stages, and this proves to be one of the most straightforward arrangements, avoiding some of the difficulties occasionally arising with directly-coupled and similar circuits.

VR1 is the volume control, operating with any unit plugged in. VR2 is a tone control, and will be found quite useful. Tr1 is a high-gain low noise amplifier, followed by the driver Tr2. Tr3/4 is the output pair, and output can be up to about ½-watt without heat sinks, or 1-watt with the sinks described. Negative feedback over driver and output stages improves quality of reproduction, and is by R10.

Battery, loudspeaker and amplifier are in a single cabinet, and the case listed is  $8 \times 5 \times 5$  in, giving a professional finish to the equipment, and taking a

3\frac{1}{2}in. loudspeaker.

#### **External Connections**

A coaxial socket is used for all inputs, with the inner lead for the audio circuit, and the outer brading for the positive line or earth return. The transistor tuners also need a negative supply of around 6V, and this is obtained by inserting a plug in the negative socket provided. This method was used instead of a multi-way connector because the negative supply is not always required. It is also cheaper.

The coaxial input and negative socket are on the amplifier panel, rather than behind, for quick chang-

ing of accessory units.

#### Construction

Constructors who feel they would welcome some general guidance should note that these details apply in a similar way to the other units, described later.

Components are mounted on a 1/16 in. thick paxolin

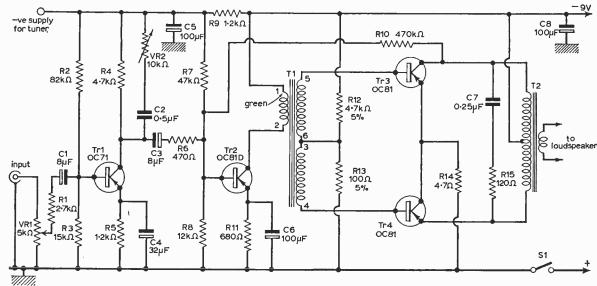


Fig. 1: Circuit of the Pyramid amplifier.

panel (Fig. 2). This is afterwards turned over, and wired on the other side. The panel should be prepared by drilling as many holes as possible, before mounting any components. Guide lines can be scratched with a straight-edge and sharp tool. A hin. or hin. drill is most suitable for the wires from resistors, capacitors and transistors. Where 6BA bolts are used for mounting and other purposes, holes can be hin. or hin. If a few holes are missed, these can be done carefully later.

The wire ends of resistors and capacitors are bent over to pass through the holes. Sharp right-angled bends should not be made immediately against the component body. Resistors and non-electrolytic capacitors (such as C7) may be placed in either way, but electrolytic capacitors (C1, C4, C5, etc.) have positive and negative ends, which must be located as shown.

In Fig. 1, all resistors except R12 and R13 are 10 per cent, so have silver bands in addition to the three bands showing the value. R12 and R13 are 5 per cent, so have gold bands.

The driver transformer T1 has a green spot to identify tag 1. Tags pass through holes, emerging as in Fig. 3. Leads soldered on hold the transformer in position. Other driver transformers should give similar results. but as leads or tags will

probably be in different positions, wiring data must then be taken from the maker's leaflet.

T2 is bolted in place. Leads from it also pass through the paxolin. Alternatives of similar type should be satisfactory. T1 and T2 should be of correct ratio and type for Tr2, and Tr3/4, with a 2/3 ohm speaker. If to hand, it is feasible to use other transistors and the transformers for them, such as OC71 driver, and 2 × OC72 output. Power output would be reduced, but should be adequate. R7, R8,

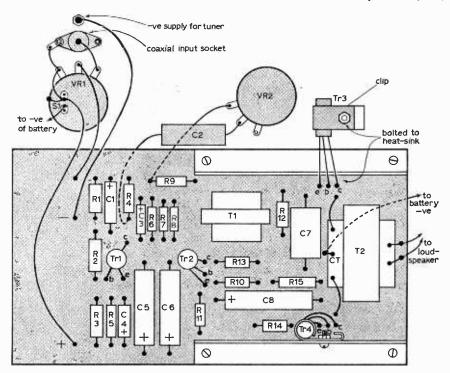


Fig. 2: Component layout.

R11, R12, R13 and R14 should then be chosen to suit the transistors.

#### **Heatsinks**

These are 16 s.w.g. aluminium, approximately 3in.  $\times$  2in., a flange being bent to bolt as in Fig. 2. Tr3 and Tr4 are placed in transistor clips, bolted to the sinks. The values given allow operation up to 500mW without heatsinks, and this is often adequate. For the maximum output of 1-watt, the sinks are essential, and R12 is  $2 \cdot 2k\Omega$ , R13 39 $\Omega$ , and R14 3·3 $\Omega$ .

#### Wiring

Underside connections, Fig. 3, are most readily made with 26 s.w.g. or similar quite thin tinned copper wire. Add insulated sleeving where necessary. Connections are soldered quickly with a correctly heated iron, and the projecting ends of resistors etc. snipped off.

If pieces of sleeving about  $\frac{3}{8}$  in. to  $\frac{1}{2}$  in. long are put on the transistor wires, this will prevent them touching each other, and hold the transistors at a suitable height. It is helpful to identify these leads with coloured sleeving. For example, red for collector, green for emitter, and base wires bare.

The transistor leads need not be cut as short as possible, and a little extra length reduces chances of overheating the transistor. These connections are soldered rapidly—it should only be necessary to apply the iron for a second or two. Remove it immediately the joint is made.

#### **Feedback**

If other transformers are used, leave the Tr3/4 end of R10 disconnected. When the amplifier is first tested, set VR1 at minimum. Temporarily take a lead from the free end of R10 to Tr3 collector. If there is a slight reduction in background noise

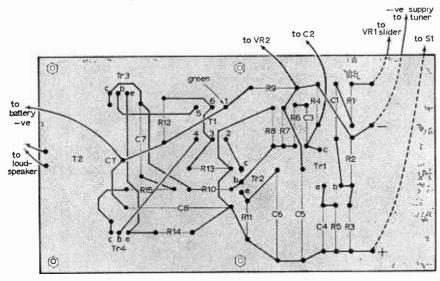


Fig. 3: Underside wiring.

#### components list

Resistors:						
R1	2·7Ω	R9	1·2kΩ			
R2	82kΩ	R10	470kΩ			
R3	15kΩ	R11	$\Omega$ 089			
R4	4·7kΩ					
R5	1·2kΩ	R13	100Ω 5%			
R7	<b>47</b> kΩ	R15	120Ω			
R8	<b>12</b> kΩ					
VR1	5kΩ log pot with s	witch				
VR2	$10k\Omega$ linear pot					
,	All ¼-watt 10% excep	ot where	e shown			
Capacitors:						
C1	8μF 6V	C5	100μF 12V			
C2	0.5µF	C6	100μF 12V			
C3	8μF 6V	C7	0·25μF			
		C8	100μF 12V			
Semico	enductors:					
Tr1	OC71					
Tr2	OC81 D					
Tr3/4	2 × OC81					
Miscellaneous:						
$\begin{array}{cccccccccccccccccccccccccccccccccccc$						

(or improvement in reproduction, if a signal is available) this is correct, and is soldered. Should oscillation occur, disconnect R10 at once and take it to Tr4 collector.

A metal panel is supplied for the case listed. This was changed to hardboard by drawing round the metal. The speaker aperture is the same diameter as the cone. The panel was given a single coating of quick drying "hammer" finish paint.

Two bolts securing the speaker also hold brackets, cut from scrap aluminium. These brackets support

the amplifier near the top of the cabinet.

VR1, the coaxial socket, and negative socket are near the Tr1 end of the amplifier, Fig. 2. They are connected as shown, the metal case or bush of VR1 being earthed to the coaxial socket and positive line, as indicated.

VR2, with C2, occupies a position on the other side of the speaker. Connections are in Fig. 3. Also connect T2 secondary to the speaker, and solder on clips for a PP9 battery. The battery rests in the bottom of the case and its clips should not touch the metal case. If the clips have bare metal, the battery can be put in a cardboard box or tray.

#### Passing R.A.E.

By reading and study of P.W. and constant listening to the intelligent discussions on the 80 metre Amateur band, I was able, without any school or correspondence course, to pass the Radio Amateur City and Guilds Certificate, and after some hard work I managed to pass the P.M.G. morse test at 15 words per minute.

How often have your readers written asking for an easy licence? Any average intelligent person can pass the necessary tests providing they are not lazy. I believe that those who complain about the technical and morse tests are too lazy to work for what they want. Possibly this is the result of the Welfare State! People get too much without any effort except making their mark if they are unable to read or write.—W. Curtis (N. Ireland).

#### Advert applause

After reading many letters of complaint that you have published in recent issues, I thought that you would like one complimentary one of your many advertisers from whom I recently made a purchase. Unfortunately the item proved faulty. This was returned by parcel post and within the space of eight days a new replacement was received by me without extra charge.

—E. Pilkington (Liverpool 14).

Correspondence regularly appears complaining of actions of firms advertising in P.W. and P.E. I would like to say that I have been well treated by advertisers with whom I have communicated, often at their expense. For example, a well known Company sent tran-

sistor circuits and data airmail at a cost of £3 though their components are not available in New Zealand.

I have just received dry reed switches eight days after posting the order—therefore it often takes longer getting sterling as we are only allowed one 5s. British P.O. a day.—K. McAllister (Auckland, 7 New Zealand).

#### Poking fun

Various correspondents have commented on the charges made by advertisers in PRACTICAL WIRELESS, some a little heated and indignant.

The feeling of being charged a lot for a very little is by no means exclusive to the radio and electronics world and indeed after reading some of the letters, your readers seem to have been treated very well compared to myself on one particular occasion.

One jet of my automatic gas stove failed to ignite. I contacted the relevant authority requesting a man to call.

Two men arrived and one poked a match stick in an obscure hole, whereupon the ignition worked perfectly. The two men were in the house for approximately thirty seconds. For this I was charged 15 shillings. They didn't even leave the match!—D. Gibson (St. Albans).

#### Local pirates?

I consider that it is time that broadcasting, for the enjoyment of millions, should be removed from the sinister hands of politicians. Every time there is a change in government our entertainment changes.

Take, for instance, the problem of local broadcasting. The present government has made it quite clear they are anti-commercials and that the BBC are the only people capable of running a radio station. This is rubbish! What about the "pirates"?

As all your readers are probably aware the propagation of the v.h.f. spectrum is very restricted, and it is quite feasible to run 20 local stations in one area without interfering with the 20 stations in another area. What a waste for us all to have to wait for the BBC to have sufficient resources to open a station in our particular area.

Why not condense the 45% of the v.h.f. band allotted to the police, and open the rest to any organisation capable of running a properly maintained station commercially?

Imagine. A choice of 20 stations, the sober British mind boggles at the thought, but it is just a matter of doing what is best for the people who, after all, elect a government, instead of obeying an old fashioned ideal.—P. Robinson (Sussex).

#### Big brother's listening

After listening on the Amateur bands from 1.8Mc/s upwards I am truly amazed at some of the antics and comments observed.

I have always understood that the granting of a licence was to responsible people for their selftraining and the advancement of their knowledge in radio.

One hears the most inane remarks at times, giggles, squeaks, and very often, particularly on 80 metres, long drawn-out conversations on things like the preparation of oriental dishes, warts—their causes and cures to name a few. Is this the experience of other s.w.l.'s?

—R. Meachim (London N.12).

More letters on page 340

A meter placed in one battery connection should show a current of about 8mA to 12mA, with no signal, or at low volume. The current rises considerably when volume is increased.

If care is taken in construction, a fault is not very likely. Should a meter immediately show a very high current, switch off at once and look for a short-circuit. If the no-signal current is low, and reproduction is distorted, this is probably due to R12 or R13.

R12 may be too high in value, or R13 too low. Also check that R14 is not, as example,  $47\Omega$  in error. On the other hand, should the no-signal current be rather high, R12 may be too low in value, or R13 too high. This depends somewhat on individual transistors, though results are generally satisfactory with 5 per cent resistors.



Books reviewed on this page are normally obtainable through any retail bookshop. In this instance, the information printed in heavy type should be quoted.

THE PHILIPS TAPE RECORDING BOOK
By Frederick Purves. Published by Focal Soundbooks. 239
pages, 7 x 4\frac{1}{2}in. Price 19s. 6d.

THIS is not a mere publicity handout; not a production devised specifically to boost the name of Philips. At least, in the hands of this experienced author it has become more than a mere exposition of Philips machines and accessories. Mr. Purves has managed to pack into this paperback what almost amounts to a treatise on tape

recording techniques.

This is a revised and enlarged version of a book that originally appeared in 1962 and was reprinted in 1965 and has now gained a complete section. It must obviously have the most appeal to Philips owners, and should provide them with much direct advice, but it is also an excellent general guide. The first chapter describes how a tape recorder works, very briefly and simply. Then the theme is developed in Chapter 2, describing the various parts, including many hidden from the ordinary user. Mechanics and electronics are treated separately. Tracks are the subject in the third, and use of the tape recorder in the fourth chapters, then we proceed to some hints and tips on practical recording.

Replay is treated as a separate subject, and the tape itself receives a chapter on its own. Likewise, editing has fairly full treatment in a twelve-page chapter, with the next section describing the development from this, some tricks one can perform

with the machine.

The tape recorder has worked its way into many spheres; industry, education, law, commerce, even politics are represented. Chapter 10 describes some of the ways the machine is used for specialist work. Accessories and connections have a very full treatment, quite rightly, and there is a short section on basic maintenance. But most important, from the engineer's and indeed the salesman's point of view is the last section of the book, which lists data on all the current and most of the latter-generation machines. This reviewer has found such collated data invaluable in his daily work: if only there were more of these collections for easy reference!

In short, this small book is likely to be of use to the average tape recorder user, the absolute beginner, and Philips owner, and even the practising engineer. What more could Mr. Purves do?—HWH.

 ≡ THE TAPE RECORDER
 ≡ By C. G. Nijsen. Published by Iliffe Books Ltd. 157 pages,
 ≡ 8½ x 5½in. Price 18s.

HIS is the second edition of a paperback that has already become a bookstall classic, even though it has to compete with many better rivals—mainly because of its reasonable price.

Written for the non-technical amateur, it attempts to pack far too much into its thirteen chapters. The inevitable result is a skimping of some information that might well have been expanded, for the sake of much that the average owner of a tape recorder may have been assumed to have learned already. Ample illustration, including 24 pages of plates and a foldout showing the tape recorder in cross-section, with a sectional view of a record/play head and some track information on the reverse, makes the book an interesting read. As may be expected from its author and his helpers, illustrations, particularly the photographs, are wholly Philips in origin.

A chapter on cassette recorders has been added since the first edition, and several of the earlier errors tidied up. In fact, the book splits fairly neatly into two sections, the last six chapters providing much useful practical advice for the newcomer to tape recording, while earlier chapters deal briefly with principles, history and rather perfunctorily with some basic theory. Interesting chapters include those on making recordings and the use of the machine in education and the study of music. Sections of earlier chapters dealing with microphones, loudspeakers, accessories and connections and, surprisingly, acoustics, will be helpful to the beginner.

Altogether, despite its ambitious over-reaching, this book is good value for its cost. Certainly, it is not beyond the understanding of the enthusiastic

amateur at whom it is aimed.-HWH.

BEGINNER'S GUIDE TO TRANSISTORS By J. A. Reddihough. Published by George Newnes Ltd. 160 pages. 7½ x 5in. Price, 15s.

HIS new addition to Newnes series of Beginner's Guides provides a good, readable, introduction to the subject of transistors, dealing with how they work, the various types available and

their applications.

The book is obviously designed for the younger reader who is becoming interested in electronics but serves equally well for any reader who wishes to brush up on his elementary theory. The author assumes that the reader starts off with negligible knowledge of the subject and accordingly begins with a brief account of the nature of electric currents before describing the principles of semiconductor devices.

Having set the scene the author goes on to discuss basic transistor circuit and characteristics then gets down to the heart of the book which delves into the applications of transistors in a.f. circuits and then r.f. circuits, the latter taking in techniques

used in television receivers.

Having covered most of the circuits used in domestic equipment such as radio sets, record players, etc., the text then branches out into the wider aspects of transistor applications. This section deals with the various types of multivibrator circuit, triggers, switches, logic circuits and so forth. The final chapters cover power supplies, integrated circuits and techniques for servicing transistorised equipment.—DC.

## repairing radio sets

#### PART 6

H. W. HELLYER

That irritating crackling, often due to a poor connection, or an inadequate aerial/earth system resulting in a loss of signal, is not one of the easiest of troubles to trace...

HE final article of this series must perforce be a kind of summing-up. Between us, Gordon J. King and I have dealt with valved and transistorised receivers from the theoretical diagnostic, servicing and general maintenance angle. Inevitably, in a series of this sort, some points have had to be skated over rather briefly. Radio servicing is a wide field, daily growing wider: we can only hope to touch upon what we consider to be the important aspects.

One of these important subjects is undoubtedly the aerial. Judging by the odd lengths of wire one finds draped around picture rails, or dangling dejectedly behind an expensive radiogram, not enough trouble is taken with this item. Purchasing a costly v.h.f. radio and attempting to receive the benefits of interference-free broadcasting on a makeshift aerial is rather like running a hot-rod on paraffin. The first step towards good reception is capturing a clean, strong signal. Proper aerial formation and siting comes high on the list of priorities.

We are familiar with the concept of wavelength and frequency. We know that the higher frequencies have shorter wavelengths, and that an aerial of "correct" length to pull in Radio 2 is king-sized compared with the multi-rod array of a BBC-2 aerial. Modern television techniques have also impressed upon us the idea of directionality, and we

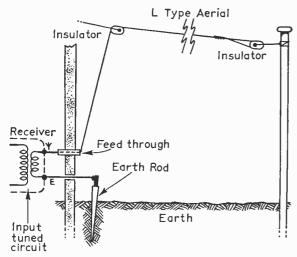
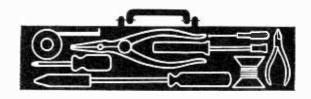


Fig. 26: Long-wire aerial is untuned, but has a natural frequency, the lowest frequency of free oscillation. But the natural wavelength is approximately 4½ times the electrical length of total aerial, including lead-in. Excess length of earth return detunes the aerial.



know that this becomes more pronounced as the frequency rises, as do the effects of reflections, or "ghosts". As regards tuning and length, long-wave aerials, and many medium-wave types, are not sufficiently critical for us to bother. In general, we used to find that the longer and higher a piece of wire was strung, the better the reception. In these days of terrible impulsive interference—when everyone's refrigerator but ours makes a switching din like an approaching earthquake—we need to take more trouble, even with the domestic radio that is never tuned to anything but ennervating "pop".

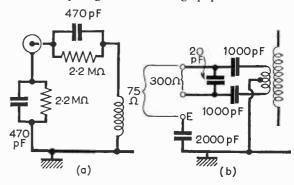


Fig. 27: (a) Isolating capacitors and discharge resistors are fitted in aerial input circuit of television receiver as protection. (b) F.M. aerial inputs may be tapped for balanced 300, or unbalanced 75 ohm feeder.

Although medium and long-wave aerials are generally "broad-tuned" for maximum pick-up of signals over a fairly wide frequency spectrum, each aerial has a "natural frequency", and if we want to get the best reception on a particular station, it is possible to improve matters by attention to this point.

Remember that signals induced into long-wire aerials are relative to earth. For best results, there must be a good earth connection. A sensitive receiver will operate quite well with no more than a short length of wire to the aerial socket—it will always be better with a properly designed aerial and earth system. Although the capacitance of the receiver, between the chassis and the earth, or via the mains supply where this type of receiver is employed, acts as an earth return, this is very inefficient. Any aerial and earth system is part of a complete circuit which includes the input coupling coil—this is as true of the tuned television input circuit as of the simple ferrite rod aerial of your portable radio. Do not overlook the input coil when fault-

finding. A loss of sensitivity can often be traced to an open-circuit or high resistance dry joint at this

point.

Where a true earth is employed, this should be as short as possible, and of low impedance. A glance at Fig. 26 will show why this is. If we regard the whole system as a tuned circuit, we see that the earth return, if long, will act as a kind of auxiliary aerial. If it is low down, as it invariably will be, the tendency to pick up electrical interference is pro-nounced. There is signal loss across the length of earth return, because of its relatively high impedance. Shortening the earth, and using thick wire, reduces this impedance, and hence the signal loss. A high impedance connection at the earth connection itself will undo all the good work that foregoing precautions have secured. It is useless to wind a piece of earth wire around the nearest water-pipe; often little better to use the convenient third pin of the house-wiring system, which is connected for best electrical safety, not necessarily for most efficient radio reception. There may be quite a long run of wire between the actual socket and the point of earthing, even though this latter should—if the Electricity Board inspectors have done their job in their usual thorough way—be an effective true earth.

Best method of earthing is undoubtedly a direct line to a rod or plate buried in the soil outside the building. Earth rods can still be obtained—even if an enquiry arouses "old-fashioned looks". An alternative is an earth connection to a rising water-main pipe, but beware of the modern plumbing systems that incorporate lengths of plastic piping. Make sure the pipe really is a water main, coming up from the ground—and never connect to a gas pipe,

or, indeed, within four to six inches of one.

If an a.c./d.c. set is being used, take great care about earth connections. As we have already noted, the chassis of such sets is connected to the neutral line of the mains supply. If the input wiring from the mains supply is reversed, or wrongly phased, it may be the "live" line that connects to the chassis. Fitting a direct earth can then result in—at the least—an almighty "bang". At worst, it will mean a windfall for your widow. If your a.c./d.c. set has no earth terminal, do not fit a direct earth to the chassis. An indirect earth, using a capacitor, of perhaps  $0.001\mu\text{F}$ , of high working voltage, is one solution, but not the best for radio reception. There is no true answer, except to do what the makers should



Sealed tuning capacitor, with access to trimmer capacitors, in a pocket portable set.

have done in the first place, i.e., fit an isolating transformer.

There is a lot of misunderstanding about the safety aspect of aerials-especially among the women-folk, who imagine that an array on the chimney attracts the lightning like a witchdoctor's fetish conjures spirits. This is only true of a poor installation. Much of the mistaken thinking comes from the use of the term "lightning conductor" for the copper rod or other gewgaw we find fitted at the highest point of many houses. A better name for such a device would be a "discharge conductor", for its job is not to capture and "conduct" lightning -even the thick copper earthing strip to which it should be fixed would rapidly disintegrate under the resulting current—but to drain away the electrical charges in a stormy atmosphere before they build up into the intensity necessary for a lightning flash. The other reason that the misconception about aerials exists is that on occasions when a house is struck by lightning, it is usually the chimney that is the striking point. This is partly because it is the highest point, but also depends on the temperature. The internal temperature of the house is generally higher than the air around. From the chimney there will exude a stream of warm air, which gathers in a cloud, becomes to some extent ionised and easily conductive and attracts the lightning by offering an easy discharge path. The nearby presence of a properly installed rooftop aerial will help disperse the ionisation of the chimney cloud and will then help prevent lightning strikes, not attract them. Tell that to the missus!

#### **AERIAL TROUBLES**

The danger of the poor installation is that the mast and fittings are not always efficiently bonded to true earth. Parasitic elements and cross-booms can be taken care of quite simply by a direct earth wire connection to the mast, using thick earth wire, (large diameter copper). Normal method of ensuring added safety, especially to a television installation (although these remarks apply equally to the v.h.f. aerials we are discussing), is the provision of discharge resistors across isolating capacitors in the aerial circuit, see Fig. 27.

For some strange reason, aerials, once erected, tend to become as neglected as an urban back-yard. Signal loss caused by corroded joints in the crossmembers of an aerial array can be severe. Aluminium makes a poor chemical ally to iron or steel, and the fixing screws often rust prematurely. The answer is an annual dismantling and treatment with protective paint—but very seldom do we find this done. At the least, the connecting points should be checked, wire ends scraped and brass or copper seating bushes polished. This takes little time and repays the atten-

tion handsomely.

Feeder cables warrant a regular inspection. Where flat twin feeder is used for v.h.f. installations, it is important that no pinching or scarring has taken place. The impedance matching (300 ohms with this type) depends on the preservation of spacing and on there being no unwanted earth contacts along the way. Much the same thing applies to coaxial, 75 ohm feeder, where pinching spoils the coaxial nature of construction, and earthing causes effective signal loops, and standing waves, which vary the required matching.

Danger points are at gutters and corners of windows, where movement due to wind and weather will cause chafing. If tile and gutter clips are not already in use, their fitting might be worth considering. On a fresh installation, or if there is sufficient spare length to modify an existing one, leave a small loop at the window entry to allow water to drip off, rather than be conducted through the entry hole to form puddles on the carpet which may get the dog a bad name. See Fig. 28.

Internal aerials hardly merit a mention. They

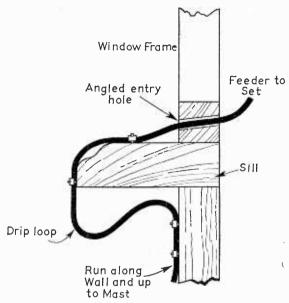


Fig. 28: When installing aerial cable, leave a drip loop outside entry point, and angle entry hole upwards, to avoid ingress of moisture.

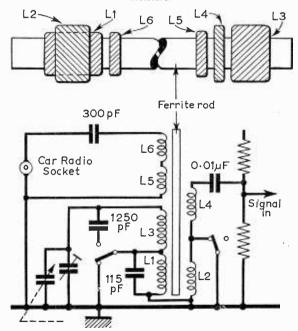


Fig. 29: Physical and theoretical layout of aerial coils as fitted in a conventional two-band transistorised radio. Note the coupling coil for car radio or external aerial connection.

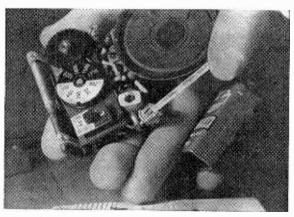
should receive occasional attention, especially at joints where movement may have occurred. And the type which is fixed inside a cabinet, such as the foil strip f.m. aerial tacked around the inside of a radiogram, or the odd sprayed screening on the backing of a radio cabinet, need hardly any attention, except to ensure that when fixing screws are removed, they will be refixed securely.

The telescopic aerial of many a transistor radio is a source of trouble and noise. This is usually because of the various methods employed to connect the coupling coil to it. Spring clips are one popular device, and the danger here is that regular use can weaken the spring. The other type of fixing depends on a screw connection, generally at the base of the rod, and looseness here can cause crackles or obscure noises, as well as detuning, even when the aerial is not apparently in use. Unfortunately, the chromed steel or brass does not readily take solder, and some form of mechanical contact is needed. It pays to make the connection as permanent as possible when reparing, with some makeshift form of Jubilee clip, or similar device. A little copper gauze clamped under a home-made loop, which is bolted together to form a clamp, with a connecting tag under the bolt, is as good a method of connecting to such aerials as the original construction.

Ferrite rod aerials are our main concern. They are the most modern form of popular aerial, and surprisingly efficient. But the fine wire of the aerial coils is vulnerable, and often difficult to solder when an accident has occurred. Quite half the "loss of signal" jobs that are brought to the workshop turn out to be fractured rods or disconnected coil connections, and some can be very tricky to trace out and repair. For this reason, I have drawn out a typical aerial circuit and its practical coil connection layout, to show the physical routing of the coil wires. Note that the long-wave coil is a pile-wound type, which may be the honeycomb variety. On this, and on the smaller, longer medium-wave coil, one end is brought out from the inside of the coil, and sometimes at the underside of the former. It is in the perverse nature of things, that a fracture will occur in this lead. A great deal of patient "fishing" and delicate work with tweezers and a small soldering iron may be needed

to rectify the fault.

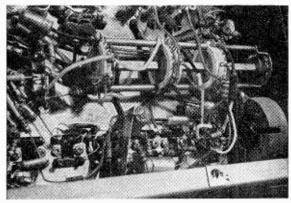
The position of the coils on the rod is important for correct tuning, and will affect the sensitivity over



View of small receiver, showing vulnerable points; aerial rod, tuning drive, slide type wavechange switch and rotary trimmer capacitor.

the appropriate band. Remember that many circuits employ both coils on long wave and only the medium wave coil on this band, so that reception on one band only can be a connection fault.

Broken rods are frequent offenders. Being of rather brittle material, ferrite rods will not withstand severe physical shock. But it should be remembered that they are non-conducting, high permeability ferrites, and not "metal" in the usual sense. This means that their continuity has to be magnetic, but not electrical, and if a clean fracture has occurred, the fault can be overcome by simply bringing the ends together again with a strong contact adhesive. There is no need to ensure a good electrical joint.



Multi-bank wavechange switch of the sealed ceramic variety. Although difficult to clean, this type does not suffer from the 'tracking' problems of the paxolin switch. Note the heavy 'flywheel' of the tuning drive in this a.m./f.m. tuner section.

Talking of electrical connections brings us logically enough to switches, without mention of which no series of this nature could be brought to a conclusion. They are the nightmare of many a service engineer, especially those mounted directly on—one might almost say "in"—a printed circuit board. Being the part of the set which receives the most frequent physical movement, it is always a likely cause of trouble. Some types of switch are so flimsy that repair work on them is practically impossible. Others still follow the robust, if clumsy-looking, style that we used to meet in the early valved superhets.

#### SWITCH CONTACTS

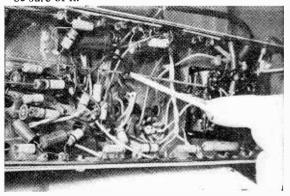
The simple type of switch consists of a bridge clamped in insulating material, sliding to connect with a pair of blades. The blades are soldered directly into the printed board; the slider rides in a bracket which is also soldered to the board. The principal cause of trouble is sideways movement of this bracket, either through damage, or by loosening of the soldered joints. This prevents the bridge from "making" cleanly and may cause it to butt against the blades, displacing them, aggravating the fault, sometimes rendering the whole switch beyond repair. Replacing one of these multi-contact slide switches is no joke. We have already dealt with the technique and need not repeat ourselves.

There should be little need, either, to dwell upon the replacement of the rotary-type switch. But a few words on prevalent faults may not come amiss. One of the bothers is a raising of the wiper section so that it forces the spring blade of a static section out of place. It needs only one clumsy twist, or a bit of grit in the switch to initiate this action. Then the spring action is lost, the switch becomes erratic, and bending the contacts back into place is a work of art. Another fault is loss of contact between the soldered section of some switch contacts and the spring blades themselves. This marriage of parts depends on a clamp rivet, and a little judicious soldering can remedy matters. But the utmost delicacy is needed.

Again, where the "ring" at the centre of some rotary switches is earthed—a common device in audio gear—and the connection has to be remade, a little blob of solder wandering past the limits of the ring can short against the stub ends of the reverse side of the switch contacts and cause intermittent faults as the switch is operated. This is a particularly difficult fault to trace unless one has met it before and is ready for it.

Switches that depend on a mechanical action, such as those with lever linkage, spring blade snap action, spring return, or a simple actuating lug, are all subject to the faults of wear. Insufficient switch travel, poor return action, intermittent contacts and other faults can result. The remedy is always to secure correct mechanical travel first, then complete the linkage and observe the switch travel before tackling the movement of the switch itself.

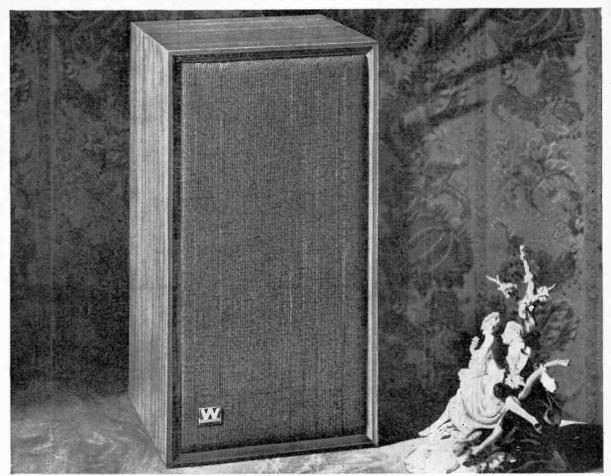
Switch cleaning we have also dealt with, but repeating that the latter-day aerosols, and modern switch-cleaners are great worry-savers can do nothing but good. Do not be tempted to a makeshift switch repair. This part takes an awful lot of punishment—be sure of it.



Earth connections are easy to trace where a common bus-bar is employed, but joints such as those indicated may be prone to looseness, and cause noise.

Another moving part that can be troublesome is the tuning drive itself. In many modern transistor radios the ganged capacitor is a sealed unit and faults here will demand replacement of the capacitor. But in others, open blades can be distorted by clumsy handling, dirt can find its way between blades, or at the contact points of the rotor wipers, causing odd noises. Routine cleaning is the answer to many of the problems.

In fact—to sum up the practical side of the whole series on radio repairs—common-sense cleaning and maintenance is more often the solution to problems than refined diagnosis and delicate fault-tracing. I feel sure Mr. King would agree with this conclusion, and join with him in hoping that the series has helped toward the easier handling of troublesome sets.



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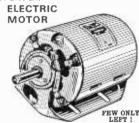
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# practically wireless HENRY commentary by

#### **Exhibitionism**

T the time of preparing this deathless epic the Radio Show looms up, now splintered into a multitude of taken-over hotels where manufacturers vie with each other to combine the intimacy of a domestic atmosphere with the opulence of high life. Sad to say, the domestic bit is too often a perch on the edge of a bathtub. between the gin bottles and the sales manager's dyspepsia tablets, where our intrepid reporter is avidly digging for exclusive news of the latest Superset contribution to "the art of listening in style".

This wish for secrecy has brought about some interesting contretemps over the years. Even now, dear old Ferrograph, sweating on the secrecy of a revolutionary new model of their "incomparable" tape recorder, have hidden it behind an advertising curtain. Only to find the gaff has been blown: first one magazine, then another, published a picture and those tantalising preliminary specifications that make enthusiasts itch to handle it.

How do you think these gems of pre-Show information come to us, Joe? Do you imagine a team of gaberdined reporters, fedoras pulled low over hawk-like features, drooping fags, bottomless gullets, etc., etc. Or has that idea of the Special Correspondent permeated yet?

Henry was once briefed to write his personal impressions on one Radio Show, with particular emphasis on its opening. Assiduously, he did his homework, reading up the BBC Yearbook, the catalogue of the three previous exhibitions, all the "top secret" handouts of the major manufacturers. On preview day, he broke all the sprint records in haring round the stands in search of titbits that other reporters may



Broke all sprint records

have missed. Privately he gloated over those layabouts in the Press Office who never ventured farther than the nearest bar.

Opening day dawned, and the queues around Earl's Court sweltered in the heat or froze in that awkward corner where the Underground tunnel disgorges. Even the passport of a Press Card did not help Henry, for Mrs. H. and one of the little aspirates had to be shepherded in. So by the time the opening ceremony began, a perspiring Henry was jostling at the back of the crowd, straining to hear the bigwigs' choice deliveries. P.A. is not, repeat not, an exhibition strongpoint.

something approaching despair, Henry repaired to the Press Room. There was almost as bad a congestion, for all the experienced technical journalists and Special Correspondents were huddled round a television set watching a closed circuit relay of the proceedings in comparative comfort. Some seemed to have nodded off to sleep. Yet next day the reports were concise and accurate. What they had not gleaned from the Press handouts they filled in with overheard snippets. Some had not stirred

from their seats since the doors opened.

At a later date, Henry slogged for a whole afternoon round an Audio Fair, snarling at intruding chambermaids who wanted to turn down the hotel bed. His piece, dotted with technical scoops, was sub-edited out of recognition by the following week, by which time Messrs Superaudio had sprung a "late late" surprise with plug-in printed boards or upside-down sync or reflex feedback or some such wonder—which, of course,



Snarling at intruding chambermaids

Henry no longer races to report exhibitions. That is a job best left to the experts. He goes to browse around the stands and get a general impression of what next season's sets will look like. He uses the excuse of a trip to Town to meet a few of his old Trade acquaintances. He might, if he is lucky, pick up a few brochures to mull over in the train on the way home.

Which is why Henry can even now write an "authoritative opinion" of the Radio Show. Whatever happens, it will be safe to say it was the same as last year.

## G.W. HARDY'S table KEYLESS PART THREE POTTABLE ORGAN

O conclude, the author discusses playing technique, and describes an audible warning device which will operate if the lid is closed, with the

organ switched on.

We will first deal with the best method of using the wands. A small piece of cloth should be kept in the stowage compartment, as the plates will need an occasional rub along them, to make sure they are clean. Also before playing rub the wand contacts on the two pieces of felt provided in the stowage compartment, one piece at each end near the front. The author prefers to use the wand with the sustaining switch in the right hand. Always make firm contact with the plates. One uses the back half of the plates mostly with the right hand, and front half with the left hand. It is best to start with some simple tune in Key C until one gets used to the instrument. Practice with expression at low volume and with left and right hand alternately. Also with the switch on Wand 2 held closed with the forefinger. Do not forget that this switch is normally open, and no sound will come from the instrument until it is closed. If for convenience in playing one wishes to change hands and the sound to remain unbroken, just place the wand on the same plate, either can be removed, without a break in sound, as long as one or the other remain in contact.

With practice the facility of playing a melody by sliding the wand contact from one plate to another can be made use of—some tunes lend themselves to this. Legato playing is to be preferred, and is done

Guiding wood block Front half of lid closes down on rear surface needle, sounding bottom C if fixed to case organ is left switched on. side, bass end. Plunger goes through larger hole in contact board between side of case and C# plate Plastic knitting-Stiff Sustain switch tags on needle freely pre-amp control panel leather moving in hole Fixing screw holes -4.5V (B2) tag on note generator panel Bottom C tag 2 screw R.H. contacts P bronze double springs biased upwards

Fig. 8: Details of the warning switch

by playing a note before removing the contact from the other plate. For instance to play a C scale in legato fashion place LH Wand on middle C, place RH Wand on D and afterwards lifting up LH Wand and placing on E, and in like manner through the scale or melody.

Now try the sustain switch, release the switch, the capacitor across it will now slowly discharge place the wand in contact with a plate there may be no sound or it may still be decaying, close the switch when it will cause the sound to quickly rise in volume. There is no contact click, but a fascinating quick build-up of the sound, release the switch and it will slowly decay, contact must remain firm with the plate all the time.

It is interesting to bear in mind that the sustain switch is still operative, even if only the LH Wand is in use. With practise many pleasing effects can be obtained, some not possible on a keyboard

instrument.

#### A warning Device

It has been decided to fit a warning device on the instrument, in case it is closed up with the switch left on, a very easy thing to do. A simple double-pole on and off switch biased off, was made up as shown in Fig. 8. P/Bronze cut to ¼in. strips was used double to increase the spring pressure. The two springs have to close on the contacts together, so a piece of stiffish leather was cut to form a nonconducting bridge across the ends of the springs. The leather has some Bostik on the soft side, and the spring ends are scraped clean and Bostik applied, on being brought together a good union is assured.

The two short pieces of 4in. square wood acting as stand-off pieces are drilled in the centre to take a screw to fix to the bottom. Before fixing, arrange a position between the side of case and the C Sharp plate for the plunger to work freely up and down, the bottom of the plunger to push on the leather. The switch is built on a small piece of the hardboard. When the switch is pressed down the sustain switch is closed and preamp operating, at the same time 4.5V is connected to a chosen plate screw under the plate board. Bottom C was chosen as being the handiest so a soldering tag was placed on the screw of that plate and nut fitted.

A small block of wood was drilled to make a sin. plastic knitting needle an easy fit. The block was fixed to the inside left of case and another larger hole drilled in the contact plate to correspond with the hole in the guiding block. Connect wires as shown, and when finished pressure on the plunger will sound bottom C loudly. Now arrange that when the front half of the cover top is closed the plunger will push hard down on the contacts, and sound

bottom C.

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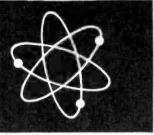
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#### UNIVERSAL CONNECTOR FOR EXPERIMENTERS

K. RAINBOW

HE idea for this unit was conceived when the author was experimenting with some audio equipment and kept finding that the screened leads to be used seldom had the right plugs to fit the equipment. The unit was designed so that a connection can be made quickly, in one of many different ways, between four plugs and four sockets. Up to four different connections can be made at one time and the unit can also be used to join leads having a different plug and socket.

The circuit is obvious from the wiring diagram (Fig. 1) and is self-explanatory. It can be expanded or reduced to the reader's requirements, in fact if an extra plug and socket are intoduced there are 45 possible different connections! The wander sockets, 1-8, are interconnected using four short lengths of p.v.c. covered wire with wander plugs on the ends. Purists (like the author) may worry about the lack of screening but in fact in the prototype no trouble at all has been experienced due to hum pick-up. All the same the leads should not be made longer than is absolutely necessary to go from socket 1 to socket 5.

#### Construction

Because the author has two left hands the unit had to be housed as simply as possible. This was achieved using pieces of hardboard nailed on to wooden ends to form a simple box. The top is drilled as in Fig. 2 or to fit the components—the dimensions are really immaterial and the unit could be made much smaller if desired.

When all the components have been fitted, the screened wire is cut into approximately lyd. lengths and the plugs are fitted. The other ends are passed through the top holes and knotted inside to secure them before wiring up.

Wiring up can be carried out with 22 s.w.g. tinned copper wire providing the connections are kept as short as possible to reduce the risk of hum pick-up. The wire can be sleeved if desired, but this is not necessary if the design of the prototype is followed. Figure 1 shows the wiring diagram of the underside. All that remains now is to test the unit by passing an audio signal and comparing the output with that given by a straight-through lead. When experimenting it is useful to remember that wander socket No. 9, which should be the black one, is connected directly to the screening.

The cost of this unit is low and depends on the proverbial junk box, but even if everything is bought new, the unit soon repays its cost in usefulness.

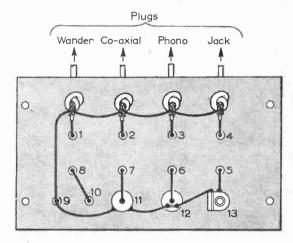


Fig. 1: Wiring diagram.

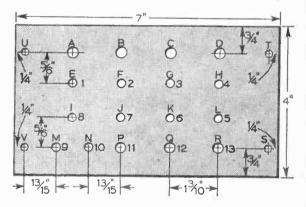


Fig. 2: Panel drilling (see text).

#### \* components list

1 jack plug; 1 jack socket; 1 phono plug; 1 phono socket; 1 coaxial plug; 1 coaxial socket; 9 red wander plugs; 9 red wander sockets; 1 black wander plug; 1 black wander socket; 4 yds. screened wire; 22 s.w.g. tinned copper wire; p.v.c. covered wire; hardboard; softwood; nails etc; 4 No. 6 woodscrews ½ in.

## REFLEX FRONT END

REFLEX circuits have a reputation for being gimmicky and unstable. This one has proved reliable in various applications and the unit described should present no difficulties provided the layout is strictly adhered to and the coil wound exactly as detailed.

The output can be fed into a conventional transistor a.f. amplifier with volume control, a four transistor push-pull type having proved adequate in sensitivity and gain. Space is available behind the panel for an additional paxolin board or veroboard on which an a.f. section can be mounted if desired.

#### Circuit Description

The aerial is fed via a 250pf trimmer capacitor to an aperiodic winding on the coil, loosely coupled to the tuned secondary winding, which in turn passes the tuned r.f. signal to the transistor base via a third inductive winding. The r.f. signal is amplified by the AF118 and then passed to the diode where the signal is demodulated and the resultant a.f. fed back to the base of the transistor via the  $0.1\mu\mathrm{F}$  capacitor.

The signal is again amplified and being a.f. passes through the 2.5mH choke and buffer capacitor to the output.

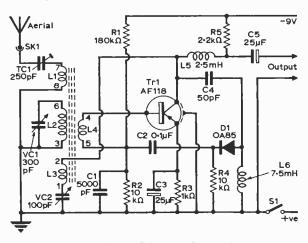
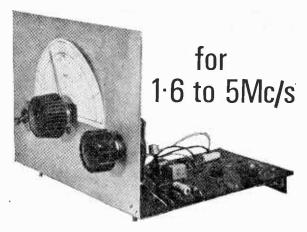


Fig. 1: Circuit of the reflex front end.

Regeneration is accomplished by a positive feed-back loop (r.f.) from the collector to the tuned circuit and controlled by the 100pF variable capacitor.

The transistor is stabilised by the potential divider resistors in the base circuit and the resistor/capacitor network in the emitter line. A  $10k\Omega$  resistor gives bias to the diode.



#### by V.S. EVANS

It will be seen that the theoretical diagram coincides closely to the layout drawing showing the reverse side.

The tuning capacitor can for convenience be one section of a two gang, or the two sections in parallel provided this equals about 300pF. The one used by the writer was a surplus 196 +110pF having a slow motion spindle, thus avoiding an expensive slow motion dial. The panel is aluminium to minimise hand capacity effects.

The coil is made from a 2in. ferrite rod fill in diameter, around which is glued a sleeve of thin card. A further sleeve line line long is made to fit loosely over the previous one and is the coil former. The clear, quick setting glue sold in tubes is best for this job. The wire used may be 26 s.w.g. enamelled (or cotton covered) and the windings should also be

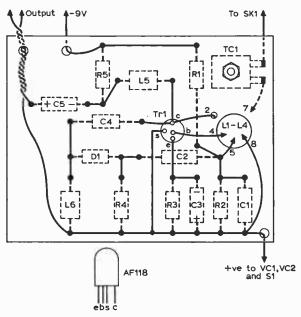


Fig. 2: Layout and wiring diagram. Components shown dotted, are on the reverse side.

### \*

### components list

### Resistors:

R1 180kΩ R4 10kΩ R2 10kΩ R5 2·2kΩ

R3  $1k\Omega$ 

all 10% 1W

### Capacitors:

C1 0·005μF

C2 0·1μF

C3 25µF 12V electrolytic

C4 50pF

C5 25μF 25V electrolytic

VC1 300pF air spaced

VC2 100pF solid dielectric

TC1 250pF compression trimmer

### Inductors:

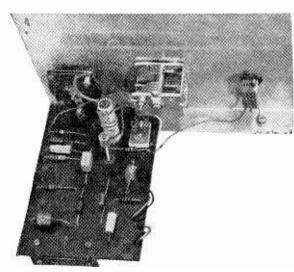
L1, L2, L3, L4 see text and Fig. 3 for winding details

L5 2.5mH r.f. choke

L6 7.5mH r.f. choke

### Miscellaneous:

8 x 6in., 18 s.w.g. aluminium; 5 x 4in.  $\frac{1}{16}$ in. paxolin; AF118 transistor; OA85 diode; S1 on/off toggle switch; battery clips; PP3 9V battery; transistor holder.



Above chassis view of the assembled receiver.

smeared with glue to keep them in position. The ends are taken through holes made in the former to the inside and brought out through the ends as per diagram.

It will be seen that leads 1 and 2 are kept away from the others as they are "lively" and carry the

r.f. feedback.

The usual precautions should be observed when soldering small components to avoid overheating. The layout ensures that all leads are kept short and direct. In most instances the actual component wire ends can be used for wiring up. Where extra lengths of wire are required it will be found that 18 s.w.g. tinned copper wire will meet all requirements. Before

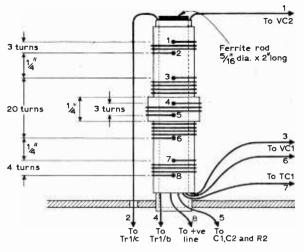


Fig. 3: Coil winding, see text for further details.

connecting the battery, make sure that all wiring is correct, as the transistor can become damaged.

Ensure that the collector is not in any way shorted to earth in connecting the regenerative capacitor. A solid dielectric type has been chosen as a protective measure, but if an air-spaced type is to hand it is as well to connect a 0.01 µF capacitor in series as an added precaution.

### **Tuning Procedure**

Tuning is best carried out by bringing the circuit to the threshold of oscillation, with the regenerative capacitor, and then rotating the tuning capacitor through the frequency range. It is important *not* to let oscillation actually take place (whistles or squeals from the loudspeaker) as in this condition the aerial is radiating and may cause interference to other listeners. The 250pF trimmer will help to match the aerial/earth system used and balance sensitivity with selectivity.

For the experimenter the range of frequencies covered may be altered to suit personal requirements by adding or subtracting to the number of turns on the main winding of the tuning coil. Up to ten turns off will bring in the 49 metre band whilst adding ten turns will bring in Radio Luxembourg.

### IMPORTANT

DON'T BUY PRACTICAL WIRELESS and then allow it to become torn and dirty. Don't search frantically through your back issues for that particular article either.

Treat yourself, and your magazines, to a *Practical Wireless* Binder and Index. A complete year's issues all in one place with an index for quick reference. The Binder is available for just 14s. 6d., and the Index costs only 1s. 6d., postage and packing included. State which volume number you want on the binder, if you don't, we'll send you a blank one.

Available from the Binding Section, George Newnes Ltd., Tower House, Southampton Street, London, W.C.2

# the Unit in the Un

OST people engaged in the design and development of all but the simplest items of electronic equipment, would agree on the advantages accruing from having the use of a good resistance substitution box.

Laboratory class R.S.B.s are expensive and will not be considered. The R.S.B.s that will be covered will prove to be of considerable use for much development, and all servicing purposes. The factors requiring consideration when drawing up the specification of an R.S.B. are: 1. Degree of accuracy required. 2. Resistance range to be covered. 3. Wattage rating required. 4. Cost in relation to 1, 2 and 3. 5. Method of selecting required resistor.

### **FACTORS**

Taking these in turn, the degree of accuracy required (factor 1), is commensurate with the type of work undertaken. Half or one per cent resistors are an expensive luxury if only routine development or servicing work is undertaken. Conversely, 10% or "open tolerance" 20% resistors are clearly useless if required for very refined work. The best tolerance for all round work, including work of an experimental nature, is almost certainly 5%, particularly in view of the fact that good quality 5% hi-stab resistors are available, at the time of writing, at most reasonable prices from a number of advertisers.

As far as factor 2 is concerned, it is probably safe to say that an R.S.B. should cover as wide a range as practicable, otherwise the work involved is likely to be curtailed. Even if immediate needs are adequately served by an R.S.B having only a limited number of resistance values, provision should certainly be made for subsequent extensions as and when required.

The usefulness of the R.S.B. can be curtailed by factor 3 if work involving medium or high currents is undertaken. The current carrying capabilities of a resistor, expressed as a wattage in terms of  $E \times I$ ,

 $I^2R$  or  $\frac{E^2}{R}$  where E, I and R are voltage, current,

and resistance respectively, must be adequate for the job in hand. Overheated resistors can suffer a resistance shift, so it is preferable to either derate the resistors in use or to select resistors of somewhat higher voltage than actually required. Unfortunately, high wattage resistors having a reasonably close tolerance (5%) are almost always wire wound, and therefore correspondingly more expensive.

All the previous three factors are involved in that of factor 4, cost. This is a problem for the individual to solve. The foregoing remarks, plus his own experience, wealth (or lack of it!) and requirements must be carefully considered and a working balance effected.

As far as factor 5 is concerned, ease of resistor

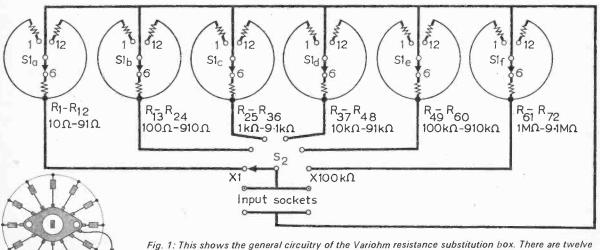
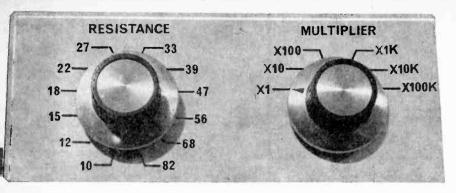


Fig. 1: This shows the general circuitry of the Variohm resistance substitution box. There are twelve resistors connected to each wafer of the switch, but only three are shown in the interests of clarity.

The small inset (left) shows the idea, however, this is only a pictorial drawing and it should not be taken that the switch looks exactly like that sketch shown in the inset.

he six rings are connected to positions 1-6 of 52



# resistance substitution box

# by H.T.KITCHEN

selection makes some form of switching mandatory. There are various possible switching combinations, For all-round versatility, the switching circuitry used in the prototype under discussion is as good as most (and better than some). This is shown in Fig. 1 from which it will be seen to consist of a six-pole twelve-way switch, and a single-pole six-way switch. Before investigating the switching further, it may prove instructive, principally for our beginner friends to

digress a little and look at "preferred values". These are values originally applied to resistors, but now being also applied to capacitors, for manufacturing convenience. There are 13 values in sequence 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82, 91. Multiples of  $\times$  10,  $\times$  100 etc, up to  $\times$  100k $\Omega$  are also available, providing resistors from  $10\Omega$ up to  $10M\Omega$  in the same sequence. Thus the first sequence starts at 10 and ends at 91, the second starts at 100

and ends at 910 and so on. Strictly speaking, as written above, the highest resistance available is  $9.1 M\Omega$  (91)  $\times$  100k $\Omega$ . A 10M $\Omega$  resistor is available, and this is usually the highest value commonly used.

### SWITCHING

To revert to the switching, the preferred values are selected by the six-pole twelve-way switch S1a-f, each of the decades being associated with each of the six poles, S2 the single-pole six-way switch being used to select the decade required. An example may perhaps clarify matters. Suppose a resistor of  $560 \, \mathrm{k}\Omega$  was required, S1 would be set to  $56\Omega$  and S2 to the  $\times$   $10 \, \mathrm{k}\Omega$  position.  $56 \, \times \, 10 \, \mathrm{k}\Omega$  gives us the required  $560 \, \mathrm{k}\Omega$  resistor.

Since there are 13 preferred values, and the most commonly used switches have a maximum of 12 positions, it is unfortunately necessary to break the sequence by dropping a value. In the prototype, for reasons that are not relevant to this article, the value dropped was "91". Readers interested enough to build this R.S.B. will have to decide for themselves which value to drop.

Constructionally, the R.S.B. is very easy to build and, with careful shopping, need not prove expen-

sive. It is contained in a home-made cabinet, to the dimensions of Fig. 2. S2 is a rotary "wave change" switch, with two poles, each of 6 ways. In order to reduce contact resistance the two halves of the switch are wired in parallel. This is by no means essential, and readers having single-pole switches having 6 or more "ways" can certainly use them.

SI is a Radiospares Maka-a-Switch having six separate single-pole twelve-way wafers. There should

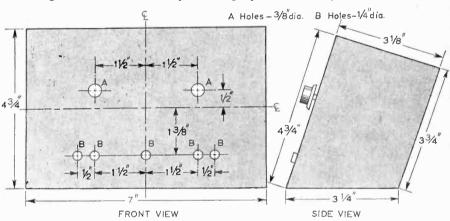


Fig.2: Dimensions of the cabinet and drilling details of the front panel.

be little difficulty experienced in wiring this up, provided the orientation of the wafers on the shaft is borne in mind, though this difficulty should not arise if the switch is mechanically assembled prior to soldering the resistors on to the wafers. It is mentioned because I myself found it more convenient to solder the resistors on to the wafers and then assemble these on to the switch shaft.

### SOCKETS

It sometimes happens that it is desirable to know the magnitude of the voltage being developed across a resistor. In experimental work, there are usually plenty of connecting leads going to the circuit being developed. It was therefore decided to add an extra pair of sockets, in parallel with the R.S.B.s input sockets, to facilitate the connecting of a voltmeter in order to measure the voltage being developed across the resistor in circuit.

As the input sockets and internal "works" are electrically "floating", it was found necessary to provide an additional earth socket in order to earth the cabinet and minimise the pickup of hum voltages when the R.S.B. was used on high impedance circuits.

# Personal RECEIVER

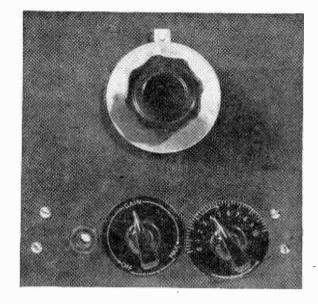
by D.V. Debbage

HE main shortwave receiver being out of commission due to an extensive rebuilding programme, a unit was required which would enable listening contact to be kept on "eighty" and top band—both phone and c.w.—but which at the same time had to be relatively simple and quick to build.

### The Circuit

The circuit finally arrived at is shown in Fig 1, the first two stages being quite conventional.

As c.w. facilities were required, initial thoughts centred on a separate b.f.o. but a little experimenting found this to be unnecessary. The heart of this little receiver being Tr3 and associated circuitry from which it will be noted that a regenerative detector is used instead of the usual separate diode.



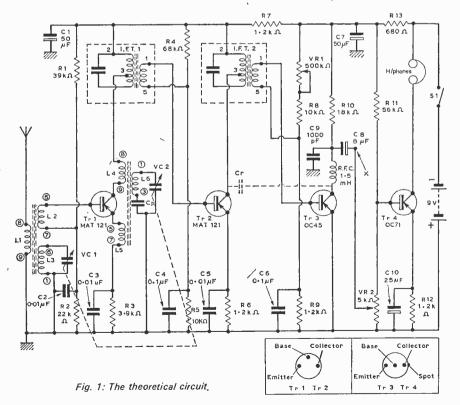
In the interest of simplicity only one i.f. amplifier is used, the loss of the usual second i.f. amplifier being compensated for by using high gain transsistors in the Tr1 and Tr2 positions, plus the judicious use of regeneration in the detector stage. Tr3 thus fulfils three requirements, these being an increase in gain, increased selectivity by narrowing the response curve of i.f.t.2, and rendering c.w. and most s.s.b. signals intelligible.

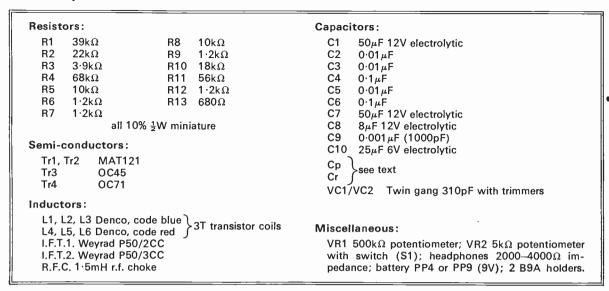
For a.m. reception VR1 is advanced to a point just short of oscillation, and for c.w. and s.s.b.

reception the detector should be just gently oscillating. Incidentally, seventeen different regenerative configurations were tried entailing six different types of transistors before a satisfactory circuit was found. In the prototype this detector stage slid gently into oscillation with VR1 two-thirds advanced.

In Fig. 1 a dotted capacitor (Čr) is shown connected between the collectors of Tr2 and Tr3. This is the regenerative feedback path and it will probably be found testing that there sufficient stray capacitance present for correct operation. However, should regeneration not take place at any setting of VR1, a small fixed capacitor in the region of 2pF can be used, or a "twisted pair" may be tried.

As stated earlier, cover was required for "eighty" and top band and the Denco Range 3 coils were used—blue for the aerial





coil and red for the oscillator, the capacitor denoted Cp in Fig. 1 being 1100pF.

There is no reason why other ranges should not be used and if this is done the value of Cp and the pin connection for connection of this capacitor can be ascertained from the leaflet supplied by the manufacturers with each coil

The i.f. transformers listed were used solely because they were to hand at the time of construction but other transistor types may equally be used. If available, a double tuned transformer in the i.f.t.1 position would be preferable.

The detector is shown feeding a single stage amplifier which is more than adequate for use with phones. If it is decided to feed a separate amplifier and loudspeaker, Tr4 and associated components may be omitted and the a.f. signal taken off at the point marked X.

Regarding the volume control VR2, this is a necessity as the audio output is controlled by this, and in no event should the regeneration control VR1 be used to control volume as sometimes suggested for circuits of this type. The function of VR1 is to keep the detector in its most sensitive state at all times.

### Construction

The prototype was built on a paxolin sheet  $6 \times 6 \times \frac{1}{16}$  in. with a similar piece bracketed on to form a front panel.

A general wiring diagram is shown in Fig. 2 but no measurements are given, as the size of individual twin gang capacitors varies somewhat but the relevant component positions and connections can readily be seen.

Although a single range is to be covered, it will be noted that the plug-in method has been used because with these coils it is

-continued on page 337

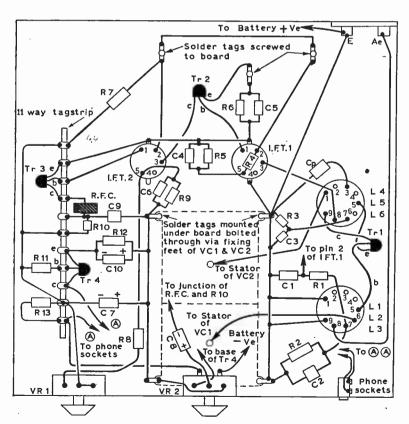


Fig. 2; Component layout and wiring on the underside.

# Five to hi-fi

### PART FIVE ..... YOUR EQUIPMENT AND YOU

IAIN SMITH

UALITY of sound is, to some extent, a personal thing. It has been said that one man's fidelity is another man's distortion. No hard and fast rules can be made as to what turntable should go with what pick-up etc. It is all really a matter of personal choice. There are some parallels to draw, however broad you may wish to space them. For instance, you would not, through sheer common sense, fit a ten-and-sixpenny crystal cartridge into a £12 pick-up arm, the reason being that the vices of the cartridge would heavily outweigh the virtues of the arm. So it is up to you and your pocket. Only a rough guidance can be given and this is what this series of articles has tried to do.

### Choice

Stereo or mono? The answer to this depends not only on your pocket but also on your circumstances. If you are living in a bed-sitting room measuring nine feet square, a stereo system could not be recommended. True there are several small stereo systems on the market but the loss of quality, which must accompany these small systems, does not make them a good investment for high fidelity fans. It should also be remembered that a mono system costing, say, £100 will usually be far superior in quality to a stereo system of comparable price.

Whether you choose stereo or mono it is recommended that you choose a stereo pick-up and if necessary suitably connect it for mono operation. If you decide on stereo, you may be wondering why you should not go out and simply buy a one piece stereogram and be done with it. There are several factors, so let us run briefly through some of them.

The main reason is that no matter how large the stereogram is and even with speakers at either end, the correct position of the speakers for stereo listening is rarely achieved. The distance between each speaker should be the same as the distance from the listener to either speaker. Another reason is that when speakers are in the same housing as the pick-up and turntable, acoustic feedback oscillation is a hazard. This is due to vibration from the speakers being fed back to the turntable and pick-up and being re-transmitted through the system. If you still decide on a stereogram, remember that, in some models, the quality of the parts that count can sometimes be questionable. Most of your money may be for the cabinet.

What equipment should you purchase? Let us assume that you are going to have a stereo system and you can budget for £100. Your dealer will recommend several different makes but here are

some pointers to bear in mind. For reasonable quality in an integrated turntable and pick-up unit the Garrard SP.25 fares well. Fit this with a Decca Deram stereo cartridge and for less than £20 plus, say, £5 for a plinth, your weakest link is off to quite a good start. If, like the majority of P.W. readers, you want to build your amplification system, then perhaps two Mullard 510s plus the Mullard stereo pre-amplifier might be considered. Several firms advertise kits for these. If you prefer to buy, then £35 will get you a Rogers Cadet Mk. 3 stereo valve amplifier. For speakers a couple of Wharfedale Super 10s in suitable enclosures, which you may or may not build yourself, and your system is ready to go. Good quality should be obtainable from this type of system.

Just one tip if you buy the Garrard SP.25. This unit has a facility whereby at the end of each record the pick-up lifts, returns to its rest and the unit switches off. This feature is not really necessary in a high fidelity unit and removal of the mechanism reduces lateral friction of the pick-up arm to one-tenth of the original value! Removal of the mechanism is a simple job and it certainly reduces record wear

### Other Considerations

Having installed your system, you will now have to make sure that you do your part to keep it at peak performance. This means, essentially, care of your records and stylus. One of the most hazardous occupations in which a gramophone record can indulge is dust collecting. There are several devices on the market for removing dust from records but prevention is better than cure. The Watts "Dust Bug" is a device with a felt roller and a brush, which traverses the record while it is playing, rather like a pick-up, collecting dust from the surface of the disc.

If your pick-up has a hydraulic raising and lowering device, then use it. It can operate the pick-up far more gently than your hand and it will protect your stylus and records. Remember, if you drop your pick-up you can damage the cartridge or, at the best, you can chip the stylus. A worn or chipped stylus will damage your records and that is a reminder to renew your stylus regularly. Do not wait until you can hear something wrong for, by then, the damage is done.

Make sure that all interconnecting leads between the various units are screened. If you are troubled by hum this can be due to "earth loops" sometimes caused by screened leads having the screens earthed at both ends. To obtain best results from speakers, try different positions, as this can sometimes improve matters. Generally, speakers perform better

when placed in the corner of the room.

With stereo systems it is important that left and right hand channels are connected correctly. It is also important to ensure that the speakers are phased correctly. That is, to ensure that both cones, when fed with the same signal, move in and out together. If they do not there will be detraction from the overall sound and a noticeable lack of "presence". One way to check is to listen to a mono recording with a good deal of bass, while reversing one pair of speaker leads. With the speakers out of phase a loss of bass will be most noticeable. When carrying out this test remember to switch the appropriate selector on your amplifier to "mono" thus paralleling both channels.

A good idea, when setting up a stereo system is to buy the Decca L.P. "How to give yourself a stereo check out". This disc has about one dozen tracks, each designed to test a different aspect of your system. Having set up your sound system, you may want to add a v.h.f./f.m. tuner in order to get high fidelity radio reception. If you have a mono system then you have two alternatives, either to build from a kit or buy. If yours is a stereo system,

then at the moment, you will have to buy.

### **Broadcasts**

Stereo transmissions are beamed on multiplex signal containing information for both left and right hand channels. How then does one aerial and one tuner give signals for each channel? To do this a stereo decoder must be used. This is connected between the tuner and stereo amplifier and splits the multiplex signal into l.h. and r.h. channel information suitable for feeding into the amplifier.

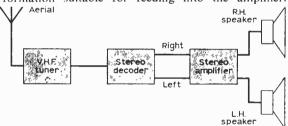


Fig. 7: Block diagram of a simple scheme for the reception of stereo broadcasts. The various units are at present available both ready built and in kit form from advertisers in P.W.

Whether you consider stereo radio worth while is another matter. The BBC transmissions are very sparse at the moment and there is little to indicate that this will change in the near future. It is, however, definitely on the way. Tuners have not been specifically dealt with in any part of this series because most people will probably wish to start with a record reproducing system. Having, I hope, given you enough information to do this, I am sure your interest, and appetite, will sharpen sufficiently to make you investigate tuners. If you feel like building one, how about starting with the transistorised switched FM tuner described in PRACTICAL WIRELESS a few months back?

This series was not meant to be a technical thesis on record reproduction, merely, as the title suggests, some general notes on the subject. I hope it has been helpful.

### PERSONAL 4 RECEIVER

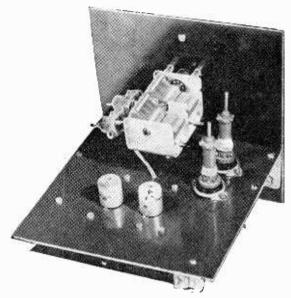
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very easy to damage the formers when directly soldering to the pins, and furthermore other ranges can be used as desired.

When all wiring is completed, check all connections carefully, and if satisfied all is in order, connect

the battery and switch-on.

The alignment procedure will not be repeated here as this has been described many times in these



Rear view of the Personal 4.

pages, also complete instructions are given in booklet D.T.B.4 available from the coil manufacturers for the modest sum of two shillings.

During alignment keep VR1 fully retarded until completion when i.f.t.2 can be peaked with VR1 advanced to the point of near-oscillation when it will be found that the sensitivity and selectivity of the unit will be extremely good for such a relatively simple receiver.

A vacancy exists on the staff of

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Experience in this field is essential. Applicants are invited to contact—The Editor,

Practical Wireless/Practical Television, Tower House, Southampton St, London W.C.2.

### AND THE NEXT, PLEASE!

To borrow the well known advertising phrase—That's Roy, that was! We had to insert a footnote to the June leader owing to the substitution of Roy Mason for Edward Short (himself only recently substituted for Anthony Wedgewood-Benn) as the resident Postmaster General. We now have to bid farewell to Mr. Mason and welcome John Stonehouse to a seat that hardly had time to get warm. We are offering no prizes for guessing who will be next month's PMG!

# Starting Next Month SPECIAL 8-PAGE PULL-OUT SUPPLEMENT

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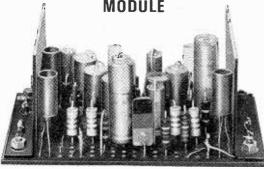
# **AUDIO**

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### WIRELESS INDEX

The index to Volume 43 of *Practical Wireless* is now available from the Post Sales Department, George Newnes Ltd., Tower House, Southampton Street, London, W.C.2.

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### FIVE STEPS TO HI-FI

In Part three, page 203, second column, 40th line. This should read . . . "where n is the ratio of the number of turns of the Primary to those of the Secondary."

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339

### Sorry

Having just noted the appearance of the June issue with my article "Experimental Transistor Millivoltmeter" pages 116-118, I append a list of errors for which I apologise. p.116 column 1, line 20... "of between say 50µA and 2mA". p.118 column 1, line 13... "of values for VR1". p.118 captions to Figs. 2, 3 and

4... "For R2 read VR1" p.118 column 2, line 7 from the bottom... "currents below 100 µA". p.118 column 2, line 2 from the bottom... "sensitivity between 50µA and 2mA".—P. Williams (Paisley, Renfrewshire).

### The other leg

Regarding Mr. G. Williams' (E.17) letter in the July issue, I wonder if he is going to publish a decoder key for his interesting information. I wonder if a binary code was used. Some economy of "bits" seemed to have accrued, but then redundant information seemed to have crept into some "words" (Perhaps there are some teething troubles in the equipment). By the way, I have NEVER been able to get my OC45's to oscillate above a few (very few) mega-hertz/cycles (envy!)—K. Smith (E.17).

I would like to comment on the subject of the OC45 transistor, owned by Mr. G. Williams, who claims in the July issue of your magazine, that it functions at 30Mc/s.

The manufacturers of this device claim an upper frequency limit of 6Mc/s, and I would like to point out to Mr. Williams that this transistor is of the alloy-junction construction. This is fabricated by alloying a metal containing large amounts of p-type impurity on to each side of a wafer of n-type semiconductor, this latter forming the base. By the very nature of this type of construction, the base region is physically rather thick, typically of the order of 0.0002 inches.

Now, it can be shown that the high-frequency cut-off, ft, is inversely proportional to the square of the base thickness. Other factors do enter into the expression, but this is the predominant term. Thus,

it can be seen that the thicker the base region, the lower will be the high-frequency cut-off, which can be shown to be no greater than 6Mc/s, for the OC45.

Transistors which do operate at very high frequencies are generally of diffused construction, where n-type and p-type impurities are diffused into an n-type collector region, for an n-p-n transistor. Base thicknesses of less than 0.00005 inches can be achieved using this technique, thereby increasing the value of f<sub>T</sub>.

If Mr. Williams does have an OC45 which operates at 30Mc/s, then I suggest either that his device is incorrectly labelled, or that he send it back to the manufacturers for further investigation; after all, he might have discovered some new effect!—C. J. Gibbins, B.Sc. (Liverpool).

### Solid status

Solid state apparatus employs transistors and diodes of silicon or germanium. The components are usually mounted on printed circuit boards and the connections are soldered. There are no valves employed and no components which are liable to deteriorate with age. The solid state circuit produces only small amounts of heat unlike many valve circuits which sometimes appear to compete with electric heaters for honours. It is however more likely to be damaged by heat than valve apparatus and should therefore never be mounted near any source of heat.

In general, therefore, a solid state apparatus will continue to work after being dropped on the floor, whilst any other type won't. This criterion however is not recommended as a practical method of differentiating circuits.—P. Hardy (Surbiton).

In reply to Mr. Tomlinson (July, 1968) I think I know the answer. Solid State. We have seen a recent example of how to mystify the uninitiated where an easily understood symbol c/s, meaning cycles per second, became Hz pronounced (by me) HURTSES in memory of childhood pains in the physics lab, where a sadistic schoolmaster demonstrated alternating current with a shocking coil.

"Several thousand volts, my boy, but it doesn't hurt because there's no current."

I think that same schoolmaster must have passed through the liquid state to the SOLID STATE and given the name in memory of himself. (Jones minor used to add "FROM THE NECK UP").-G. C. Turner (East Grinstead).

In response to W. J. Tomlinson's query on the term "Solid State" (July issue). This term is given to a piece of electronic equipment which has no moving parts, such as relays. etc. For example: One could call an audio amplifier a solid state device, in the way that it is purely electronic.—G. Drinkwater (Manchester).

### Those 'B' licensees

With reference to Mr. E. Mason's letter in the July issue I feel I must protest. I do not claim the ability to read the PMG's mind but perhaps I might suggest a reason for the lowering of B licensees permits to two metres.

Like the high frequency bands the two metre, seventy centimetre and twenty-three centimetre bands are harmonically related. Thus standard seventy centimetre practise is the use of a low powered two metre rig to drive a tripler-amplifier. with a similar arrangement for 23cm. Since the facility for twometre operation therefore already exists why not allow Sound B licensees to use the band? The same conditions do not apply to the four metre or lower bands since no such harmonic relationships apply.

Mr. Mason appears to believe B licensees inferior to their Sound A counterparts. Why? We pass the same written examination, our technical competence is the same or, in some spheres, better. What, may I venture to ask, is the difference between a good B licensee and his A counterpart operating in the same mode on 70cm or above? I cannot speak for all Sound B amateurs but I, and I know of a number like myself, have never wanted to operate on anything other than Sound B territory. In addition my interest is the phase to amplitude s.s.b. technique, so why take the morse test?—R. G. Mannas, G8BIK (London, N.22).

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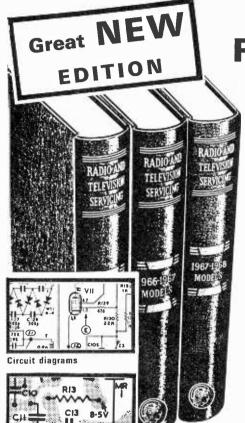
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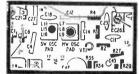
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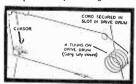
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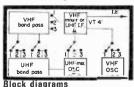
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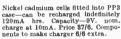
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A.S.CARPENTER G3TYJ

HIS easily duplicated low power modulator was initially constructed for use with the "Ten-Fifty" Transmitter, details of which were given in an earlier issue\* of this magazine. When c.w. is the preferred operating mode the "Ten-Fifty" Tx is complete in itself signalwise but a modulator may be considered a useful accessory with which to indulge in phone QSO's occasionally. The modulator to be described may be used with rigs other than the "Ten-Fifty" for it is an entirely self-contained unit. An "all new" construction cost is approximately £12 for some 10 hours' work.

An interesting feature of the QRP modulator is the use of the miniature audio valve, type Brimar ELL80, physically no larger than the more familiar EL84. The ELL80 comprises of two output pentodes in a single envelope which, when worked in Class AB1 push-pull can deliver up to 9W of audio power

when suitably fed. Quite clearly 9W is inadequate power with which to effectively plate/screen modulate a 50W r.f. signal but it is sufficient to so modulate r.f. inputs up to 20W. When it is required to modulate a r.f. input of 50W or so the transmitter may be cathode modulated, when some 5-6W of audio is sufficient. In

some cases it may be possible to inject the modulation at the transmitter key socket and this has been done with the "Ten-Fifty" Tx. The modulator has also been used to cathode modulate a DX-40 transmitter although in this case a separate injection socket had to be fitted.

Top Band transmitters may be plate/screen modulated by the unit of course, therefore it is but a case of deciding exactly what is required and then suitably connecting the multi-radio modulation trans-

former accordingly.

The modulation transformer used in the prototype unit is a Woden type UM1 but the smaller physically and less expensive item—the Woden UM0 is also suitable. The UM0 may be run up to 10W but the UM1 can accommodate 60W. Such transformers can often be obtained inexpensively from amateur stations gone s.s.b.—or QRT!

Plate/screen modulation is superior to cathode modulation and is also less finicky to set up; cathode modulation on the other hand is much less demand-

\*PRACTICAL WIRELESS, January 1967.

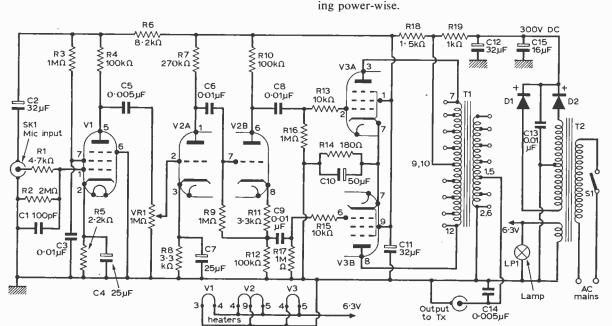


Fig. 1: Theoretical circuit of the low power modulator.

The complete circuit of the low power modulator may be seen in Fig. 1. The signal from a crystal microphone may be introduced at socket SK1 when V1 operates as a simple pentode connected speech amplifier feeding V2A, the gain control, VR1, being used between stages.

The anti-phase voltages necessary to drive the output pentodes are derived from V2B, the required voltages being developed across load resistors R10 and R12 since it is appreciated that the ends of these items remote from the valve are "earthy" to a.c. in both cases due to C11. Valve V2B functions therefore as a conventional phase-slitter and resistors R9 and R11 are merely leak and bias items respectively. Little benefit is derived from by-passing R11 with a capacitor.

The output pair function in Class AB1 and a potential of 9V positive with respect to chassis appears at the common cathode pin 7. The measured no-signal current of the cathode circuit was in the prototype 37mA and is within the maker's rating.

It should be noted that the parasitic "stopper" resistors R13 and R15 must on no account be omitted and one side of each should be connected direct to the appropriate grid tag.

The ELL80 anode-anode load required is 11,000 ohms, the secondary load (cathode) impedance being some  $300\Omega$  for the 6146 in the associated transmitter and connections to the modulation transformer are indicated in the circuit diagram. Since the prototype unit is used for cathode injection one side of the secondary winding is shown chassis connected but if plate/screen modulation is desired such a connection must be removed. Connections to T1 are then varied to suit the required load conditions and details of Woden modulation transformer connections for various loads may be found by referring to the manufacturer's instructions. In cathode connection some slight tendency to "downward" modulation has been noted at times but is not considered excessive.

Due to the meagre current drain of the low power modulator a comparatively small mains transformer may be used and a 60mA rating for the secondary is adequate. Used for full wave rectification a pair of BY100 silicon diodes are excellent and occupy little space. If a valve

rectifier is preferred—and adequate mounting space is available—the EZ80 is ideal. Fitment of a small 2.5V flashlight bulb in the centre-tap lead of the secondary winding to chassis may be a well worth-while inclusion as a protection device particularly when diodes are used.

The chassis dimensions used are given in Fig. 2 where a single section of 12 x 18in. 16 s.w.g. aluminium is bent to provide two mounting shelves of dissimilar height this being done deliberately to prevent the modulation transformer from dwarfing the valves. A single 8in. square section of 16 s.w.g. aluminium eventually bolts to the chassis underside flanges and confers both rigidity and component protection. The chassis front flange is allowed to slope backwards slightly and this not only improves the appearance but also simplifies operation of the gain control.

### ★ components list

Resisto	ors:		
R1	$4.7k\Omega$	R11	3·3kΩ
R2	$2M\Omega$	R12	100kΩ 5%
R3	$1M\Omega$	R13	10kΩ
R4	100k $\Omega$	R14	$180\Omega$
R5	2·2kΩ	R15	10kΩ
R6	8·2kΩ	R16	$1M\Omega$
R7	270kΩ	R17	$1M\Omega$
R8	3·3kΩ	R18	1·5kΩ 1W
R9	$1M\Omega$	R19	1kΩ 1W
R10	100kΩ 5%	VR1	$1 M\Omega$ log.
	All res	sistors 10%	łW

### Capacitors:

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- C2 32μF 275V
- C3 0·01μF ceramic
- C4 25μF 12V
- C5 0.005μF ceramic
- C6 0·01μF ceramic
- C7 25μF 12V
- C8  $0.01 \mu F$  ceramic
- C9 0·01 μF ceramic
- C10  $50\mu F 50V$  C13  $0.01\mu F$  paper 1kV
- C11  $32\mu F 350V$  C14  $0.005\mu F$  ceramic
- C12 32µF 350V C15 16µF 450V

### Transformers:

- T1 Woden UM1 or UM0
- T2 A.C. Mains input. Secs. 250–0–250V, 60mA 6·3V, 2A

### Valves:

V1-6AM6, V2-12AX7, V3-ELL80

### Diodes:

D1, D2-BY100

### Miscellaneous:

Chassis (see text), valveholders B7G (1); B9A (2); knobs (1); lamp and lens (1); coaxial socket; cable; wire; etc.

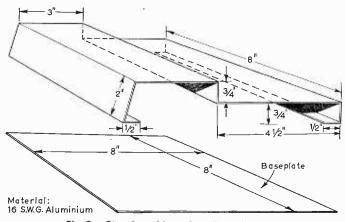


Fig. 2: Chassis and baseplate dimensions.

—continued on page 396

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# ON THE SHORT WAVES

### MONTHLY NEWS FOR DX LISTENERS

### THE BROADCAST BANDS by

by CHRISTOPHER DANPURE

Desn't time fly, it only seems last week the summer schedules started, now they have only a few weeks to go, as on September 1st most International Shortwave radio stations will change to their autumn or spring schedules depending which side of the Equator they are, so most information in this column is liable to change on that date. In next month's issue I hope to list some of these schedules for various stations. Now here are this month's propagation predictions.

West Africa: 0800-1600 25, 21, 17 and 15Mc/s; 1600-1800 25, 21, 17, 15 and 11Mc/s; 1800-2000 25, 21, 17, 15, 11, 9 and 7Mc/s; 2000-2200 21, 17, 15, 11, 9, 7, 6 and 5Mc/s; -2200-0200 17, 15, 11, 9, 7, 6 and 5Mc/s; 0200-0400 15, 11, 9, 7, 6, 5 and 4Mc/s; 0400-0600 15, 11, 9, 7, 6 and 5Mc/s; 0600-0800 21, 17, 15,

11 and 9Mc/s.

South Africa: 0800–1400 25 and 21Mc/s; 1400–1600 25, 21 and 17Mc/s; 1600–1800 25, 21, 17 and 15Mc/s; 1800–2000 25, 21, 17, 15, 11 and 9Mc/s; 2000–2200 21, 17, 15, 11, 9, 7 and 6Mc/s; 2200–2400 17, 15, 11, 9, 7 and 6Mc/s; 2400–0200 15, 11, 9, 7 and 6Mc/s; 0200–0400 11, 9 and 7Mc/s; 0400–0600 11 and 9Mc/s; 0600–0800 21, 17 and 15Mc/s.

East Africa: 0800–1200 25, 21 and 17Mc/s; 1200–1600 25, 21, 17 and 15Mc/s; 1600–1800 25, 21, 17, 15 and 11Mc/s; 1800–2000 21, 17, 15, 11 and 9Mc/s; 2000–2200 17, 15, 11, 9 and 7Mc/s; 2200–0200 15, 11, 9, 7 and 6Mc/s; 0200–0400 15, 11, 9 and 7Mc/s; 0400–0600 17, 15 and 11Mc/s; 0600–0800 21, 17 and 15Mc/s.

South Asia: 0600–1200 21, 17 and 15Mc/s; 1200–1600 21, 17, 15 and 11Mc/s; 1600–1800 21, 17, 15, 11 and 9Mc/s; 1800–2000 17, 15, 11, 9, 7, 6 and 5Mc/s; 2000–2200 15, 11, 9, 7, 6 and 5Mc/s; 2200–0200 11, 9, 7, 6 and 5Mc/s; 0200–0400 11, 9 and 7Mc/s; 0400–

0600 17, 15 and 11 Mc/s.

South East Asia: 0600–1000 21Mc/s only; 1000–1200 21 and 17Mc/s; 1200–1400 21, 17 and 15Mc/s; 1400–1600 21, 17, 15 and 11Mc/s; 1600–1800 21, 17, 15, 11 and 9Mc/s; 1800–2000 17, 15, 11, 9 and 7Mc/s; 2000–2200 15, 11, 9, 7 and 6Mc/s; 2200–2400 11, 9 and 7Mc/s; 2400–0200 11Mc/s only; 0200–0400 15Mc/s only; 0400–0600 17 and 15Mc/s.

North East Asia: 0600-1600 17 and 15Mc/s; 1600-1800 15 and 11Mc/s; 1800-0400 11Mc/s only; 0400-

0600 15Mc/s.

Australia via Asia: 0600–1000 21Mc/s; 1000–1200 17Mc/s only; 1200–1600 15Mc/s; 1600–1800 11Mc/s only; 1800–2000 11 and 9Mc/s; 2000–2200 11, 9 and 7Mc/s; 2200–2400 11 and 9Mc/s; 2400–0600 circuit closed.

West Coast, South America (North of Chile): 1200–1800 21 Mc/s; 1800–2000 21 and 17Mc/s; 2000–2200 21, 17 and 15Mc/s; 2200–2400 17, 15 and 11Mc/s; 2400–0400 15, 11 and 9Mc/s; 0400–0600 15 and 11Mc/s; 0600–0800 15 and 11Mc/s; 0800–1200 15Mc/s only.

### **NORTH AMERICA**

Canada: R. Canada is now using 17,820, 15,325 and 11,720 for its French transmission from 1315–1345 not 21,595, 17,820 and 11,720 as in last month's column, 21,595, 17,820 and 15,325 used from 1345–1830.

U.S.A.: R. New York, Worldwide (Station WNYW) had dropped 15,405 in favour of 15,440 for its beam to Europe from 1930–2200. Also, has a new transmission to Latin America in English. From 2200–2345 on 21,525 and 17,845, 2345–2400 on 17,845, 2400–0100 on 17,845 and 15,215.

### **AFRICA**

Ethiopia: R. Stn. ETLF, Radio Voice of the Gospel, Addis Ababa, is giving strong signals in its Farzi programme from 1600–1625 on 15,400, the programme is repeated from 1630–1655 on 11,935. The station's intermission signal is the "Elizabethan Serenade" followed by four notes on a drum.

### **EUROPE**

Federal Republic, Germany: R. Deutsche Welle has altered the frequency for its German programmes to East Asia, Australia and New Zealand from 0645–0940, was using 21,580, 15,205 and 11,795, now dropped 21,580 for new 9,650 beamed across Gt. Britain, Atlantic and Pacific Oceans.

Holland: The Happy Station Programme on Sunday mornings from 1030–1150 has dropped 11,730 for 5,980, still using 9,715 and 6,020. In November this year the Happy Station will celebrate 40 years on the air. A special Happy Station QSL card is being issued for the occasion for reports on Happy Station transmissions only.

Monte Carlo: Trans World Radio now uses 7,285 for its transmissions in English starting SU-FR at 0625, on Saturdays at 0610. The Sunday afternoon transmission is now aired from 1500–1615 on 7,255.

### **ASIA**

Malaysia: The BBC Far Eastern Relay station at Tebrau has now dropped 15,435 for 15,350 with 7.5kW for 1345-1815 beamed to South Asia.

Pakistan: R. Pakistan, Karachi, is now giving fair signals on new 15,330 from 1845–1930 in Turkish and English to the Middle East and from 1945–2030 in English to British Isles and Europe.

### **AUSTRALASIA**

New Zealand: R. New Zealand is now on the following schedule until September 1st. English to the Pacific Isles 1700-1945 on 9,520 and 6,080; 2000-0545 on 15,110; English to Australia 2000-2230 on 11,705; 2245-0545 on 17,770. To the Pacific Isles in English, Samoan and Niuean 0600-0800 daily on 9,540 and 6,080 0800-0845. Weekdays only in English and Rarotongan on 9,540 and 6,080, 0800-0845 Sundays only in English on 9,540.

IRST comment this month is that you'll be hearing those callsigns on c.w. signing a bit faster if they are G's. The maximum speed for sending station identification in morse is now 20 w.p.m. and not 12 as it used to be.

A very thin month for the l.f. enthusiasts. Hardly a soul mentioned any frequency lower than 14Mc/s, and the sparse DX down there seems to be even sparser when I've listened too. Twenty metres has proved its usual interesting self but definitely the Band of the Month award must go to 15 metres. There have been some very good openings and these have varied, some midday, and some later after 2100.

unearthly hours have logged some goodies.

Ten metres has been rather spasmodic, although it's usually possible to hear W stations on sideband at 5 and 9 plus, any time that this segment is stirring. When it has been open, though, ten has provided some very FB DX.

Even the early birds who raise their weary bodies at

One or two v.h.f./u.h.f. fiends have queried the fact that no 144Mc/s logs and above appear. Well, what about it all you G8's. Is two metres really worth

it, or is it just a quiet version of topband?

Just a reminder that evening classes usually begin in September. Have you checked if there is an R.A.E. course at your local institute? Many organisations run courses for R.A.E., both the written and c.w., but you should give the relevant body a ring. There might be a number of people who would like an R.A.E. course, but if the local authority don't know of this need, then they will not run a course. Go on, get on the land line now and have an a.f. QSO.

Little birds have whispered the following in my ear (left one actually). There's an HL9 either US or UF on at noon most days. Listen for s.s.b. just above 14,200kc/s. Just below this and about the same time, listen for VK9RJ. My best for the month

—on 40, VS6DO, and on 15, 9M2CP.

**P.** Pollak reckons that JX might well be the prefix for Jan Meyen Island. He heard JX6RL on twenty, anyone think different or shall I put it to the vote?

### LOGS

About 90 per cent of the logs this month were for fifteen metres. Hardly anyone listened below twenty metres, and only one or two scanned ten.

G. Haslip (Sussex), 19 set, 100ft. long wire, found 80 mostly full of Europeans (EU's), although one or two interesting calls came through, like GC3KEV and VS5RCS.

Someone who really claims an accolade for persistent listening in the face of heavy QRM. Logs

received for 80, 40 and 15 metres.

F. McVerry (Lanarkshire), R1155, 75ft. end fed is the man who did it. Scalps taken from—DJ9RY/P, LA1H, SMØLK/P/Ø, W1MFY, WA1IRN, WA2BEE, WA4YHA, ZC4RB, 4X4UF, 9M2NF. Which is pretty good for 80 metres s.s.b. On 40 s.s.b. Francis managed to hook—CN8AW, CN8BV, CN8FF, EA6BG, EA8EX, EA8FF, GC3GS, HB9RZ, JW7YF, LA2PH/MM (near Capetown), PAØDX/M, PY1CBS, PY2NE, PY4BLR, PY6HL, PY7AKW,

TJ1AL, TU2AK, TU2BQ, VE1APB, VO1AK, VU2OLK, WA3GTW, W4AVO, WB4IDW, W5FXZ, WA7IFM/MM, WA8ACQ, WØLLV, YV1BI, ZC4RB, ZS1JA, 4U1ITU, and all that was on forty. I daren't tell you what he heard on fifteen!

**D.** Grant (Kent) has just acquired a B40 receiver. He says it's working FB and the xtal is still very accurate. He also reckons that all bands are active with 10 mainly short skip with the odd South American station popping up. Fifteen is very good around 2000-2200hrs, while EU's clutter up twenty making good DX logging difficult. With a half-wave dipole at 25ft. Douglas heard these on fifteen metres s.s.b.—CE3DR, CR6EX, CX7BW, EA8BQ, EA8FF, EL4JY, ET3UN, HI8XCC, JA1PYP (running 20 watts). KG4AM, KG6ALY, KP4APQ, KV4CF, KZ5DK, LU3AX, MP4BGU, OA4OS, OD5FM, PJ3UM, PY1JZ, PY7LAV, SVØWM, VR4CR, YV1TP, ZC4MO, ZC4RB, ZD8Z, ZD9BE (Tristan da Cunha), ZS6DP, 4X4SO, 5H3JR, 9G1GD, 9K2BB, 9Q5DG, 9U5CR, 9U5SK.

### **PICKINGS**

C. Morgan (Northumberland), 10 transistor s/het, 120ft. long wire, 15 metres a.m.—CE7DW, CR6GA, CR7CY, CR8AH, FG7XW, JA5AKY, PZ1CP, TN8DR, LLANKAR, WUNDER, TW8DR, TW8DR, LLANKAR, WUNDER, TW8DR, LLANKAR, WUNDER, TW8DR, LLANKAR, WUNDER, LLAN

TN8DB, UA9KAB, VU2DKZ, 7XØVP.

P. Pollak (London), 840A, PR30, 60 ft. end fed, 15 metres s.s.b.—CEØAE, CN8BB, CR6GA, HS3DV, K5HYB, K6URK, KV4AD, KZ5NG, LU1DAB, MP4BDE, OA4UZ, PY6NG, TU2BQ, VK2NN, VQ9DH, W6USG, ZD8JW, ZD9BE, 4X4GV, 5H3JL, 9Q5ML.

M. Pasek (Notts), QP166/HRO, 150ft. long wire, 15 metres s.s.b.—HS3BA, ITØARI, JA4OK, KG4AM, KP4BCL, KZ5HL, K2UZU/P/KG6, MP4BGR, OH2BR/MM (near New York), VK9WD, VU2DKZ, VU2OLK, XW8BX, ZD8Z, 9M2NF, 9V1OC.

R. Dinning (Ayrshire), HA350/PR30, 252ft. end fed, 20 metres s.s.b.—AP2MR, CT2AW, DU2FU, EA6BQ, HV3SJ, HKØBKW, LU9DJW, ÖX3CJ, SMØCH, TF2WFI, TN8CS, UN1AB, VE8RCS, VP8CG, 5H3KJ, 5Z4KL, 9K2BJ, 9Y4VT. Robert queries NU2OF and assures that the call sign was read correctly.

G. Richards (Isle of Wight), GC-1U, 60ft. Vee end fed, 10 metres s.s.b.—CR6GS, CR7IC, EL2AK, EL2X, LU6DRB, LU7MAL, OD5FA, VQ9V, YA5RG, YV1EL, ZC4RB, 4X4VB, 5H3KL, 9J2DT, 9J2WR, 9L1KZ, 9Q5HU, 9Q5RV, 9V1NY.

### CONTESTS

Busy month for contests. September 7th—8th, DARC phone contest: 7th—8th, V.H.F. N.F.D.; 7th—8th, VU/4S7 c.w. contest; 14th—15th, VU/4S7 phone contest; 15th, 3·5Mc/s field day; 21st—22nd, SSA c.w. contest; 22nd, National D/F hunt; 28th—29th, SSA phone contest.

If you live in GM land, don't forget the Scottish Mobile Rally at the Cartland Bridge Hotel in Lanarkshire. Early October (2nd—5th) is the date for the R.S.G.B. Exhibition at the New Horticultural

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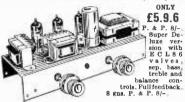
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Load lines were first used with valves, but they are equally applicable to transistors, for manufacturers' published graphs of both devices are static, that is they are compiled without anode or collector loads, so that anode or collector current variations do not produce anode or collector voltage variations. In practice, a valve or transistor cannot amplify unless it is coupled to a load, and to produce dynamic characteristics when so

valve or transistor cannot amplify unless it is coupled to a load, and to produce dynamic characteristics when so used would require separate curves for each value of load resistor. This is clearly impossible, but by the use of load lines equalling the value of

the load and drawn across a family of output characteristics, the transistor or valve currents and voltages can be determined for any input

signal.

Let's take a look first at Fig. 1 which shows a family of Va/Ia characteristics; the output characteristics of a typical power pentode. Each curve or parameter indicates the valve's anode current as anode voltage is raised from zero for different values of grid bias increased in 2V steps from 0 to —14V.

The recommended anode load for this type of valve is around  $4{,}000\Omega$ , the exact value depending on the output wattage required and the degree of distortion that can be accepted. However, if we take it to be just  $4k\Omega$ , then with a power supply of 240V, anode voltage will also be 240V at current cut-off. A dot can then be made on the horizontal Va axis to indicate this point.

At the other extreme, if it was possible to develop all the supply voltage across the load resistor instead of across the valve, current would be  $I = \frac{240}{4,000}$  Amps or 60mA. A dot marked on the vertical Ia

scale at this point will then provide the other extremity for the load line which can then be drawn across the graph.

In similar fashion, load lines for other values of load and/or supply voltage can be drawn. It will be seen that the higher the value of load resistor the more horizontal the load line becomes and that equal value load lines irrespective of supply voltage run parallel to each other. Having drawn the load line, the points where it intersects the individual curves will then indicate the actual anode voltage and current for that particular value of load and grid bias. Even more important, it will show up the disparities in the increases of anode current produced by regular decreases in grid voltage. Ideally, equal increases and decreases of bias from the no-signal or quiescent operating point should produce equal increases and decreases in anode current. Unfortunately this is never so, for the spacing between the curves for regular decreases in bias tends to increase towards zero. This means that from any practical no-signal biasing point, anode current will increase more for a fixed reduction in bias than it will decrease for an equal increase in bias. However, by careful choice of the standing bias and anode load, this disparity can be reduced to small proportions.

### **Output Versus Distortion**

The quickest way is simply to place the edge of a ruler across such a family of output curves and note where its placing, pivoted from the h.t. supply value, cuts the parameters with the most uniform results. The no-signal biasing point would then be in the centre of such a line and of course must be greater than the peak value of the input signal to prevent driving the grid positive on peak positive half-cycles. Having then

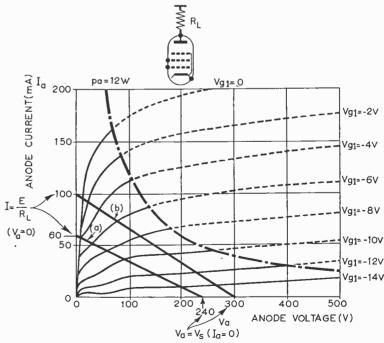


Fig. 1: Load lines drawn across anode characteristic curves of a power pentode. (a) is for a 4000 $\Omega$  load with a 240V h.t. supply, (b) is for a 3,000 $\Omega$  load with a 300V h.t. supply.

drawn a line which satisfies these requirements, its value can be determined by simple application of Ohm's law. The value of h.t. supply divided by the current indicated by the load line equals the value of optimum load. By moving the ruler over the graph it will become evident that higher outputs can be obtained at the price of greater distortion. In fact many manufacturers quote various anode loads for various power outputs with the associated maximum second harmonic distortion content. Some manufacturers issue graphs for output valves plotting the variation of anode load on one axis against the percentage of har-

monic distortion on the other axis for different values of grid bias. However, by an extended application of the load line as demonstrated in Fig. 2, the variations in output current and voltage graphically demonstrate the distortion imposed on a pure sine wave input due to this irregularity in Va/Ia curve spacing. One half-cycle becomes compressed and the other extended when related to the input sine wave.

This distortion is termed "second harmonic distortion" because it exactly duplicates the waveform of any frequency upon which is superimposed a smaller amplitude signal at twice the frequency (second harmonic) and as shown in Fig. 3. Such distortion where one half-cycle is greater than the original while the other halfcycle is relatively smaller largely cancels out in pushpull outputs circuits, but third and other odd harmonics cannot be eliminated

in this way and can only be reduced by negative feedback.

The distorted waveform shown in Fig. 3 is obtained by the algebraic addition of the two pure sine waves which means simply that similar terms are added while opposing terms are subtracted. In this illustration, excursions of the second harmonic above the zero line are added to similar excursions above zero of the fundamental, but subtracted from the fundamental below the zero line. Thus the harmonic peaks with the peak of the fundamental during the latter's first half-cycle to increase the total amplitude of the resultant, while it peaks in anti-phase with the peak of the fundamental in the second half-cycle to decrease the total amplitude by a similar amount. Furthermore, this phasing in and out mis-shapes the resultant as well as altering its amplitude, so it no longer represents a sine wave.

This distortion being caused by the non-uniformity of anode current variations instigated by the input signal is clearly indicated by the ratio of AO to OB on the load line (Fig. 2). These A and B limits are set by the peak amplitude of the input signal. If the grid input was less than 4V peak they would be closer set and have less disparity, but taking them as they stand, they indicate the maximum current changes through and voltage changes across a  $10k\Omega$  load resistor.

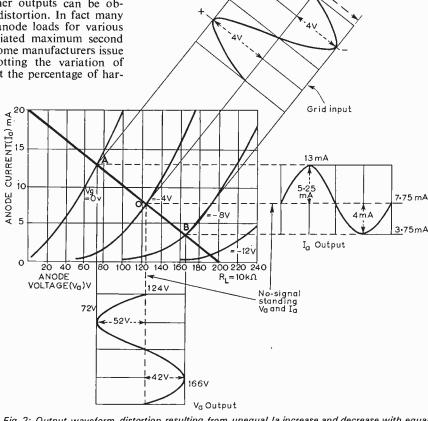


Fig. 2: Output waveform distortion resulting from unequal Ia increase and decrease with equal Vg+or-changes. Such distortion duplicates the effect of a superimposed signal at double the frequency (second harmonic) as shown in figure 3.

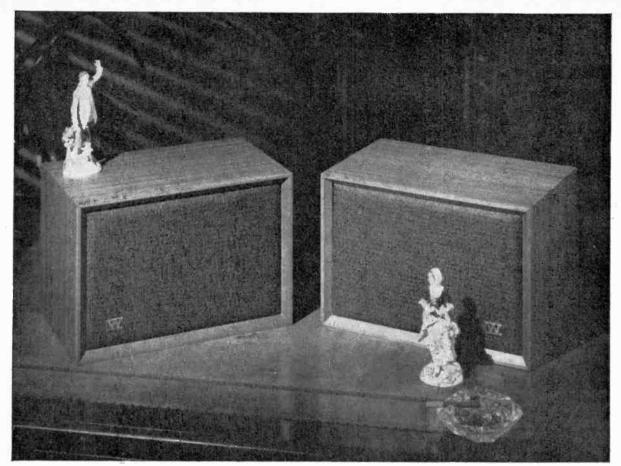
The generally accepted figure for maximum tolerable second harmonic distortion is 5% or 0.05 of the output. Therefore if we accept these figures as a maximum, the ratio AO/OB should not exceed 1.22. Ideally of course the ratio should be unity. This figure is obtained as follows: Where x equals the decimal percentage of second harmonic distortion,  $\frac{AO}{OB} = \frac{1+2x}{1-2x}$  assuming x

to be 0.05, i.e. 5%, then
$$\frac{AO}{OB} = \frac{1 + (2 \times 0.05)}{1 - (2 \times 0.05)} = \frac{1 + 0.1}{1 - 0.1} = \frac{1.1}{0.9} = 1.22.$$

Again referring to Fig. 2 it is obvious that the AO/OB ratio could be improved towards unity by increasing the supply voltage and/or reducing the value of load resistor. In the interests of clarity only 4 Vg parameters are shown, but it is evident that by careful choice of load resistor and grid bias—the latter rather less than the figure shown, an almost unity AO/OB ratio could be obtained.

If regarded purely as a voltage amplifier, the stage gain A, represented by the ratio of anode voltage swing to grid input swing will equal:

$$\frac{\text{Va (max)} - \text{Va (min)}}{\text{Vg (peak-to-peak)}} = \frac{166 - 72}{8} = \frac{94}{8} = 11.75$$
(see Fig. 2). With such purposely exaggerated second



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harmonic distortion as is present in the output of this valve, giving a 52V decrease but only a 42V increase from the quiescent value of 124V, it is vital to take the overall peak-to-peak variation to compare it with the overall input figure. When assessing power output directly from load line intersections across the graph, the peak values indicated must be converted to r.m.s. values because wattage is always expressed in these terms.

### **Calculating Power Output**

Directly calculating valve or transistor power output is often considered rather difficult—but the same Ohm's law rules apply as when calculating the output of a generator or the power dissipated in a resistor. The only difficulties that arise are due to the necessity of separating the a.c. content of the output—the real power from the steady d.c. component on which it is impressed. It can be calculated by I2R, V2/R or VI where I and V are in r.m.s. values and represent the true a.c. content of the output. To convert the peak-to-peak currents and voltages indicated on the graph to r.m.s. values, each value must be divided by 2 to determine the peak value, and then by  $\sqrt{2}$  to convert to the r.m.s. figure. For finding output wattage from the VI formula therefore,  $W = \frac{Va(max) - Va(min) \times Ia(max) - Ia(min)}{2}$  $2 \times \sqrt{2}$  $\times 2 \times \sqrt{2}$ 

As  $2 \times \sqrt{2} \times \sqrt{2} \times 2 = 2 \times \sqrt{2^2} \times 2$  or 8, this formula can be simplified to

 $W = \frac{Va(max) - Va(min) \times Ia (max) - Ia(min)}{8}$ 

With transistors in the usual common emitter mode Vc and Ic replaces Va and Ia.

Having found maximum and minimum current values, it then becomes an interesting exercise to directly measure the second harmonic distortion in the output by the following formula, which is equally applicable to valves or transistors, Io in both instances equalling the no-signal standing current. Second harmonic distortion

 $\frac{I(\max) + I(\min)}{2} - Io$ 

I(max) — I(min)

Even the simplest set of curves provide a wealth of information. For instance from the graphs of Figs. 1 and 5, it is quickly possible to find the a.c. resistance of the valves by dividing a small anode voltage variation, on the linear section of the characteristic, by the resulting variation in anode current. If the current change is expressed in milliamps, the a.c. resistance will be in

kilohms. Ra  $= \frac{\mathrm{dVa}}{\mathrm{dIa}}$ . The prefix "d" indicates "change in". Similarly  $\mu$ 

can be obtained by noting what Va change is necessary to alter Ia to the same extent as a change in Vg with Va constant. Having then obtained both Ra and  $\mu$ , mutual conductance

 $g_m$  can be found by  $g_m = \frac{\mu}{Ra}$  and keeping Ra in kilohms,  $g_m$  is in mA/V.

Figure 4, which shows output characteristics of a common emitter amplifier enable the current gain β of the transistor employed to be

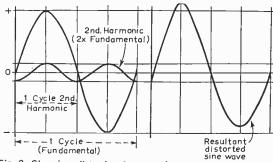


Fig. 3: Showing distortion impressed on sine-wave by second harmonic signal. Such distortion can be produced by a valve or transistor when equal increase and decrease in Vg or Ib, do not produce equal increases in Ia and Ic respectively.

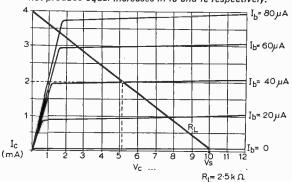


Fig. 4:  $2\cdot 5k\Omega$  load line across a family of typical transistor common-emitter output characteristics. A standing lb of  $40\mu$ A provides a good working point due to lb changes producing virtually equal lc changes. Standing lc would then be 2mA with Vc just over 5V. The graph also shows how  $\beta$  can be determined by the change in lc due to change in lb with Vc constant. In this example with Vc at 5V, a  $20\mu$ A lb change from 20 to  $40\mu$ A increases lc by closely 1mA.

Ib change from 20 to 40 $\mu$ A increases Ic by closely 1mA·  $\beta \text{ Or Hfe} = \frac{\triangle /c}{\triangle /b} = \frac{1,000 \mu A \text{ (1mA)}}{20 \mu A} = 50$ 

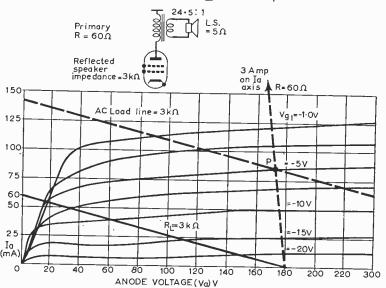


Fig. 5: When a load is transformer coupled, a load line equal to Rp provides the working point on any selected Vg curve. Load lines of equal value having the same angle to the base, a further load line equal to the effective impedance is then drawn through point "P" parallel with one of similar value. This "A.C." or "dynamic" load line shows the la Va excursions on signal input.

readily determined,  $\beta$  or Hfe= $\frac{dIc}{dlb}$  or the change in

collector current brought about by a change in base current with collector voltage constant. Such factors and indeed all others are most readily taken from graphs and once one becomes accustomed to "reading" them, can be executed very quickly. However—to keep

strictly to load lines.

Up till now, it has been assumed that the resistive load was directly connected in the anode lead of a valve or the collector circuit of a transistor, but most output stages, unlike voltage amplifying stages, employ a transformer coupled load for optimum power matching. This slightly complicates the issue of making a load line since the d.c. value of the transformer primary may be only about  $50\Omega$  while the actual load presented to the valve or transistor by normal transformer action may be several thousands of ohms. Clearly the load line must equal the Rp value for static d.c. currents, but must equal the impedance effectively presented to the signal by the transformer coupled load.

First step, therefore, is to draw a load line in the usual manner to equal the transformer's primary resistance. However, this value being so low, it will be impossible to actually terminate this line on the vertical Ia or Ic axis, as it would be several amperes, and without unduly extending the graph, must be drawn as shown in Fig. 5. The next move is to mark the most suitable or selected Vg curve where this load line intersects. The "a.c." load line must also then be drawn through this point but at an angle equal to that required by the value

of the a.c. load.

If for instance the a.c. load is  $3k\Omega$ , as load lines of equal value have the same angle to the horizontal axis, draw this  $3k\Omega$  "a.c." load line parallel to a  $3k\Omega$  "d.c." load line, drawn as a guide on the same graph. This

"a.c." load line will now indicate how the input signal varies output voltages and currents as did the conventional load line when the resistive load was directly connected to the valve or transistor.

Two facts about transformer coupled loads will now be apparent. First, the increase in Va or Vc obtained by eliminating the considerable voltage drop across the directly connected load reduces the degree of distortion introduced, while secondly, Va or Vc can now instantaneously exceed the Vs figure.

This brief outline then indicates that load lines are extremely useful and deceptively simple devices for really getting to know the functioning of valves and transistors under working conditions, and their uses can generally be summarised as follows:

- 1. Visual and rapid indication of the optimum load and best biasing point to provide maximum output with minimum distortion.
- 2. Permits direct measurement of second harmonic distortion.
- Graphically indicates the voltage gain or power output of any stage.

So for a really informative exercise, mark out the load lines for various values of load across a set of transistor or valve output characteristics, and determine the best values for the greatest output compatible with acceptable distortion. The latter of course is achieved by selecting a bias point 1b for transistors, Vg for valves, which enables equal increases and decreases in signal input to produce equal but larger increases and decreases in collector or anode current, as the case may be. Then measure the stage voltage gain or power output obtained. With output pentodes especially, it is interesting to compare your own assessment of optimum load with the maker's recommended figure.

### QRP MODULATOR —continued from page 344

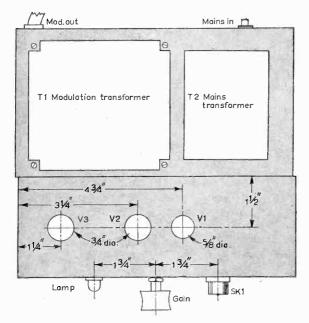
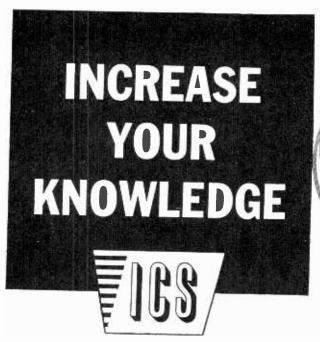


Fig. 3: Above chassis layout and dimensions.

Preliminary tests consist of checking for h.t. chassis short-circuits and measuring voltages and currents throughout the unit during which time the gain control should be set at minimum and a pair of high impedance headphones connected to the output. A voltmeter connected across the full modulation transformer primary winding should ideally read zero on its 0-10VDC range if both valve sections of the ELL80 are drawing equal currents. When a crystal microphone is connected to SK1 it will be but necessary to rotate VR1 very slightly before room noises are reproduced; speech should sound crisp and there should be no traces of instability or oscillation. Thereafter it is merely a case of bringing the modulator into general use on the lines mentioned earlier and some "on the air" tests carried out taking care not to allow over-modulation to occur.

The low power modulator is certainly a useful item but if when construction is complete everyone has gone s.s.b. and there is no one left to talk to, do not despair! By omitting VI, and associated items, connecting SKI to the top of VRI via a  $0.01\mu F$  capacitor and a  $1M\Omega$  resistor, fitting in place of TI a decent push/pull speaker transformer plus a little degenerative feedback from its secondary to V2A, an excellent little 8W audio amplifier will result. Fed by a crystal pick-up gramophone records will be reproducible with a high degree of fidelity and more particularly so if a little tone compensation is added!



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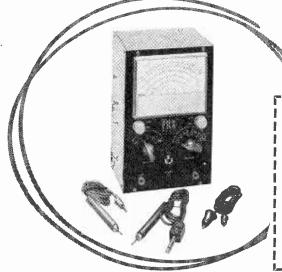
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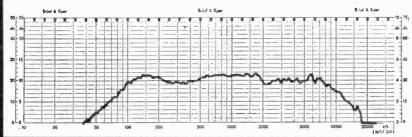


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### SINCLAIR MICROMATIC

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- FANTASTIC POWER
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PZ.4 des

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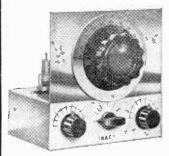
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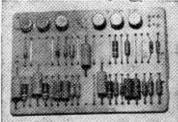
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101"	77"	1	3	0	1	8	0	1	11	9
121"	37"		17	9	1	1	9	1	4	9
121"	54"	1	2	3	. 1	7	0	1	10	6
121"	87"	1	8	9	1	14	6	1	18	9
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PRACTICAL WIRELESS, SEPTEMBER 1968

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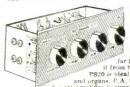
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