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Intゃrleaved and Imopregnated. aries $200-230-\% \sqrt{2} 0$. in $\mathrm{c} / \mathrm{s}$. Screened T\&P SHIRGEIDFID DIKOP IHIKOCGII $\begin{array}{llll}250-0-250 v . & 70 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a}, 0-5-6.3 \mathrm{v} .2 \mathrm{a} & 17 / 9 \\ 350-0-350 \mathrm{v} . & 80 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a} . & 0-5-6.3 \mathrm{v}, 2 \mathrm{a} & 19 / 9\end{array}$ $350-0-350 \mathrm{v}, ~$
$250 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a}, 0-5-6.3 \mathrm{v}$,
$250-250 \mathrm{v}, 100 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a}, 6.3 \mathrm{v} .1 \mathrm{a}$ $250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .3 .5 \mathrm{a}, \mathrm{C} . \mathrm{T}$ $250-0-250 \mathrm{v}, 100 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 6-5-6.3 \mathrm{v}$. 3 a Mullard 510 Amplifer $300-0-300 \mathrm{v}$. $100 \mathrm{~mA}, 6.3 \mathrm{v}$. 4 a $300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a}, 0-5-6.3 \mathrm{v}$ 3่ $350-0-350 \mathrm{v}, 150 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 0-5-6.3 \mathrm{v}, 3 \mathrm{a}$ FITLISEIHROIDED IIPRIGHT $250-0-250 \mathrm{v} .60 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a}, 0-5-6.3 \mathrm{v} .2 \mathrm{a}$ Midget type $2 \nmid \times 3 \times 3$ 3n. $250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a}, 0-5-6.3 \mathrm{v} .3 \mathrm{a}$ $300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 5 \mathrm{v}, 3 \mathrm{a}$. $300-0-300 \mathrm{v}, 130 \mathrm{~mA} .6 .3 \mathrm{v}$. 4 a ,
1 a . for Mullard Amplifler $350-0-350 \mathrm{v}, 100 \mathrm{~mA}, 6.3 \mathrm{v}$. $4 \mathrm{a}, 0-5-6.3 \mathrm{v}$. $3 \mathrm{a} \quad 38 / 9$
$350-0-350 \mathrm{v} .150 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 4 \mathrm{a}, 0-5-6.3 \mathrm{v}$. 3 a

FILLY SHIROTHFIN (oontinued)-$425-0-425 v .200 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, \mathrm{C} . \mathrm{T}, 5 \mathrm{v}, 3 \mathrm{a}$
$425-0-425 \mathrm{v} .200 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{C}, \mathrm{C} . \mathrm{T}, 6.3 \mathrm{v}$. $425-0-425 \mathrm{~V} .200 \mathrm{~mA}$
$4 \mathrm{a}, \mathrm{C} . \mathrm{T}, 5 \mathrm{~V} .3 \mathrm{a}$
$450-0,450 \mathrm{v}, 250 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{n}, \mathrm{C} . \mathrm{T}, 5 \mathrm{v} .3 \mathrm{a}$ Midget Battery Pentode 66.1 for 3S 4
small Pentode, $5,0000 \Omega$ to $3 n$
Small Pentode, $7 / 8,000 \Omega$ to $3 \Omega$.
Standard Pentode $5,000 \Omega$ to $3 \Omega$ Standard Pentode 7,000 s to $3 \Omega$ $10,000 \Omega$ to $3 \Omega$
Push-Pull 8 watts, ELSA or 6V6 to 3 ä or matched to $15 \Omega$
Push-Full 10-12 watts to match 6 V 6 or EL84 to 3-5-8 to $15 \Omega$
Following types for 3 and $15 \Omega$ speaker Push-Pull 10-12 watts 6VF or FIS 84 Push-Pull 15-18 watts. 6Lゃ, KT66 Push-Pull Mullard 510 UTHra Linear Push-Pull 20 watts, sectionally wound 6L, KTG6, EL34. etc.

MTHGET MAANS Primeries $200-250 \quad \square$ $50 \mathrm{c} / \mathrm{s} .250 \mathrm{v} .60 \mathrm{~mA}, 6,3 \mathrm{v}, 2 \mathrm{a} . \quad . \quad \ldots 11 / 9$ Both above size $24 \times 25 \times 241 n$.
FILAMENT THRANSFORMERS
All with $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$ primaries 6.3 v . $1.5 a, 5 / 9 ; 6.3 v, 2 a, 7 / 6 ; 12 v .1 a$, r7/11; 6.3v 3a. 8/11; 6.3v. 6a. 17/6; 12v. 1.5a, twice $17 / 6$.
SMOOTHING CHOKES
$150 \mathrm{~mA}, 7-10 \mathrm{H}, 250 \mathrm{ohms}$
$100 \mathrm{~mA}, 10 \mathrm{H}, 200 \mathrm{ohms}$
$80 \mathrm{~mA} .10 \mathrm{H}, 350 \mathrm{ohms}$
$60 \mathrm{~mA} .10 \mathrm{H}, 400 \mathrm{ohms}$
$60 \mathrm{~mA} .10 \mathrm{H}, 400 \mathrm{ohms} . . . \quad . \quad 5 / 8$
CHARGERE THANSFUTRMEIRS*
All with 200-230-250v. 50 c/s Primatles:
$0-9-15 \mathrm{v}$. $14 \mathrm{a}, 12 / 9 ; 0-9-15 \mathrm{v}$ a $14 / \mathrm{g}: 0-9-15 \mathrm{~g}$ $0-9-15 \mathrm{v}, 1 \frac{1}{2}, 12 / 9 ; 0-9-15 \mathrm{v}$ 2a, 14/9; 0-9-15v $3 \mathrm{a}, 16 / 9 ; 0-9-15 \mathrm{v}, 5 \mathrm{a}, 19 / 9: 0-9-15 \mathrm{v}, 6 \mathrm{a}, 23 / 9$
$0-9-15 \mathrm{v}, 8 \mathrm{a}, 28 / 9$. -9-15V. 8a, 28/8.
AUTO (Step up/step down) TRANS. $\begin{array}{ll}0-110 / 120-230 / 250 v . ~ 50-80 ~ w a t t s, ~ 13 / 9 ; ~ & 250\end{array}$ watts. 49/9; 150 watts, $27 / 9$.
MICREPIIONS THANSFOIRMERES 120:1 high grade, clamped. 8/9.

BRADFORD, BRISTOL, BIRMINGHAM, DERBY, DARLINGTON, EDINBURGH, GLASGOW, HULL,

| Brand new individually checked and guaranteed |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  | $\begin{aligned} & 0,16 \\ & 8,16 \end{aligned}$ |  |
|  |  |  |  | 250 |  | ${ }_{5}^{5}$ | $8 / 8 / 8$ |  | 8 |  |  |  |  |
|  |  | ${ }_{\text {EFY }}^{\text {EFY }}$ | KTZ6s |  | (ex |  | ${ }^{\text {P }}$ |  |  | (12AX | ${ }_{\text {\% }}^{\text {is }}$ | $8$ | 为 |
| ALbo 3/- |  |  |  |  | Yis |  | 88 |  |  |  |  |  | 9001 $81 / 6$ |
| (ex | ${ }_{\text {E1 }}$ |  | 11810 |  |  | ${ }^{818} 8$ | 4, |  | 818. |  |  |  | Heus |
|  | ${ }^{\text {Exill }}$ |  |  |  | -0.1 |  | \% 81 |  |  |  |  |  |  |
|  | EsA | EL336 5\% | ${ }_{\text {MLL }}^{\text {M }}$ |  |  |  | 101 |  |  |  |  | \% |  |
|  | ${ }^{\text {B }}$ EA | EL38 ${ }^{\text {ELi }}$ | , | OA |  |  | ${ }^{8 / 8}$ |  |  |  |  |  |  |
|  | (eater | ${ }^{\text {ELLA1 }}$ | ${ }^{\text {Jata }}$ |  | 8\% | ${ }^{\text {6AKE }}$ | 5. |  |  |  |  | ${ }^{236}$ |  |
|  | 128 |  |  | 15\% | ${ }^{76}$ |  | 80 |  |  |  |  | 70\%- |  |
| ${ }_{\text {a }}^{\text {AU }}$ | - | $8 / 8$ | O24 | 855\% |  | ${ }_{64 \mathrm{Mb}}$ | 2/8 | ${ }^{\text {¢LL }}$ | 56 | ${ }^{12887}$ |  | 8\% |  |
| 88 | - | ELs5 |  |  | ${ }_{8} 8$ \% |  | 1. | , | 6/9 | ${ }^{12887}{ }^{\text {a }}$ |  | 8 8- |  |
| 876 | E Ebrs3 | ${ }_{\text {ELPS }}^{\text {ELPI }}$ | Pcess 717. | ${ }^{180 \%}$ | ${ }^{C 8} 840$ | ${ }^{8.406}$ | \% | ${ }_{\text {big }}^{\text {Gig }}$ | 8 |  |  | 100 |  |
| 25i- | - | em81 | 8 | 81 - | N43 |  |  | ${ }_{68}^{68}$ |  |  |  | 8, |  |
| 201/ | - | ${ }_{\text {EMA84 }}^{\text {EM8 }}$ |  | ${ }_{8} 8$. | ${ }_{\text {1R4 }}^{1 \times 1}$ |  | , 11. |  |  | ${ }_{10}^{15 \mathrm{D} 2}$ | ${ }_{\text {8 }}^{2401}$ |  |  |
|  | - ECO | ${ }_{\text {Exs }}$ | PCL | ${ }_{1 \%}^{1 \%}$ | ${ }_{185}^{185}$ | ${ }^{\text {bax }}$ | \% |  |  | 19883 | ${ }_{805}^{803}$ | ${ }^{20}$ |  |
|  | - ECus |  | PCL |  | 6 |  |  |  |  | 1997 | ${ }_{808}^{887}$ |  |  |
| \% | - ECOB33 | ${ }_{\text {Exyb }}^{\text {Exy }}$ | ${ }_{\text {PCL }}$ |  | ${ }_{2 A}^{1 T 4}$ | ${ }_{\text {bab }}$ | 2/6 |  |  | 1941 | ${ }^{13}$ | 17/6 |  |
|  | EC | Ez41 | PrN4 |  | ${ }^{2826}$ |  |  |  |  | ${ }_{2084}^{2084}$ |  | ${ }^{3515}$ 30:- |  |
|  | - | 8/6 |  | U0285 ${ }^{\text {8/6 }}$ |  |  |  |  |  |  |  | 80\% |  |
|  | ${ }^{\text {ECH4 }}$ |  | \|lol |  |  | 6BW6 |  |  |  | ${ }^{25858}$ | ${ }^{832}$ | 455: |  |
|  |  |  | ${ }^{\text {Pl83 }}$ |  | ${ }^{2 C 45}$ | ${ }_{6659}^{60}$ |  |  |  | ${ }_{25869 \mathrm{~T}}$ | ${ }_{837} 8$ | 50. | $\mathrm{Cr9}^{812}$ |
| ${ }_{\text {D }}$ | EC |  | PLP |  | ll |  |  | 6VE |  |  |  |  |  |
|  |  |  |  | ULL41 | ${ }_{3 \times 2}^{2021} 85$ | ¢C66 |  |  |  |  |  | $14 /$ |  |
|  |  |  |  | Uub |  | 6cha | $1 / 8$ | 6, | 16 | 3 |  |  |  |
|  |  | ${ }_{6}^{6234} 420$ | ${ }_{\text {PX }}$ | ${ }_{\text {UY92 }}$ |  |  | 81: |  |  |  |  |  |  |
|  | (1) | ${ }^{1869}$ |  |  | ${ }^{87 / 1}$ |  |  |  |  |  | 988 |  |  |
| 5\% | - |  | ${ }_{\text {Pry }}^{\text {PY } 8_{8 / 2}}$ | ${ }^{11607}$ | ${ }_{\text {3B7 }}^{\text {sid }}$ |  | 810 | ${ }^{887}$ |  |  | ${ }^{616}$ |  | 3A/B sol- |
| - | \% | H12 | ${ }_{\text {PY }}^{\text {PY }}$ |  | ckiol |  |  | ${ }_{\text {cks }}$ | ${ }_{\%}^{10 \%}$ |  |  | \%\% |  |
| \%. | - |  | P21.3 |  | ${ }_{3529}{ }^{29} 9000$ | ${ }^{61}$ | $4{ }^{18}$ | ${ }^{718}$ | /3/ |  | \% |  |  |
| 8/8 |  |  |  |  |  |  |  |  |  |  |  |  |  |
| lit |  | KTsO |  |  | 9 | dFs | 1 |  |  |  |  |  | . |
|  |  |  |  |  |  |  |  |  |  |  |  |  | \% |



MARCONI COMMUNICATION RECEIVER CR 150/2 Frequency coverage $1.5-22 \mathrm{Mc} / \mathrm{s}$ in 4 bands. Two IF's $1 \mathrm{ss} 1,600 \mathrm{kc} / \mathrm{s}$, $2 \mathrm{nd} 463 \mathrm{kc} / \mathrm{s}$. Image signal protecting over 40 dB up to $30 \mathrm{Me} / \mathrm{s}$ and calibration (bulte-in calibrator). Stabilisacaition of supply and semperature compensation. Electrical and mechanical bandspread. Metering and visual tuning indicator. Bandpass from $100 \mathrm{cs} /$ to $10 \mathrm{kc} / \mathrm{s}$ in 5 stages. Acoustic filter associated with $100 \mathrm{c} / \mathrm{s}$. Bandpass position for CW recep. tion. Facilities for diversity reception. Excellent checked condition 635 . Mains P.S.U. by P.C. Radio 64.10.0. Carr. 301. SEALED HI-SPEED SIEMENS RELAYS $1700 \times 1700$ ohms 151 . P. \& P. 116. P.C. RADIO'S mains PSU for above 901 . TF 995A-1. Range $1.5-220 \mathrm{Mc} / \mathrm{s}$ in 5 bands. Output voltage variable in 1 dB steps from $1 \mu \mathrm{~V}$ to 100 mV . Carrier may be unmodulated, frequency modulated, amplitude modulared, or simultaneously frequency and amplitude modulated. Frequency deviacion is variable from $25-600 \mathrm{ke} / \mathrm{s}$ for FM and modulation (internal or external) up to $50 \times$ for A.M. Crystal calibrator for checking frequencies incorporated. Incremental frequency control. Fully checked and guaranteed. Price i88.15.0. Carriage in No. ig SET 17/6 each. P. \& P. 3/.
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Ditto but with built-in power supply for 210.250 y A.C. $\$ 9.19 .6$. Carriage 15 \%.
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EVERSHED MEGGER CIRCUIT TESTER. 2 ranges 0 to 1.0000, $100 \Omega$ to 200,000 2. Wish test leads, leasher carrying case. Tested $\mathbf{4} 4.19 .6$. P. \& P. 316 $32 / 44 \mathrm{ft}$. AERIALS each consisting of cen 3ft. Zin. dia. subular serew-in sections. 14ft. (7 secrion) whip zerial with adaptor to fit the fin. rod, insulated base, stay place and stay assemblies. pegs, reamer. hammer etc. Absolutely brand new and complate, ready to erect, in canvas bag 62.19.6. P. \& P. 1016.

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PANEL METERS (round) $\begin{array}{lll}0.50 \text { microamps } & 21^{\prime \prime} & \text { D.C. } \\ 0.100 \text { microamps } & 2^{* *} & \text { D.C. }\end{array}$ 0-100 microamps 0-200 mieroamps $0-500$ microamps
$0-500$ microamps $0.1 \mathrm{~mA}^{*}$
$0-1 \mathrm{~mA}$
$0-1 \mathrm{~mA}$
0.5 mA
$150-0.1,500 \mathrm{~mA}$
0.500 mV
$0-5 \mathrm{~V}$
0.15 V
0.50 V
$0-150 \mathrm{~V}$
$0-500 \mathrm{~V}$ (shunt)
$0-10 \mathrm{kV}$
"Weston", as es "S" meter.
"Weston", as usually, used in H.R.O.

#   

NEW COMMUNICATION RECEIVERS
 MODEL HE30 Covers range from $540 \mathrm{kc} / \mathrm{s}$. to $30 \mathrm{Mc} / \mathrm{s}$. Ham Band is provided with a scale for direct reading,
and can also be band and can also be band spread. Facilities: A.N.L., A.V.C and
M.V.C. Q Multiplier M.V.C. Q Multiplier
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Avaiable in lift lorm, 25 GiNs. Carr, and Pack. fret. II.P. Terms (Rfady Buit): £6.13.0 deposit and 11 monthiy patymente of £2.16.0.
MODEL HE40 covers medium wave band and 1.6-4.4 $\mathrm{Mc} / \mathrm{s} . \mathrm{A}^{4.5-11.0 \mathrm{Mo} / \mathrm{s} \text {. } 11.0-30.0 \mathrm{Mc} / \mathrm{s} \text {. in separate switched band }}$ spread ranges. Controls include B.F.O. Sensitivity, A.N.L. Receiver-Stand-by-Switch. Tone Switch, S-Meter. For 2001 250V. A.C./D.C. Internal 1000 and telescopic antennae fitted. Size $13 k \times 82 \times 54 \mathrm{in}$. Instruction manual included. No Kits available.
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plier circuits. Improved A.N., and voltage regulated powerplier circuits. Improved A.N.I, and voltage regulated power-
pack. "S" meter, band spread on amateur bands. lange illuminpack. "S" meter, band spread on amateur bands. large illuminspeaker and phones. Valve line-up: $4 \times 6 \mathrm{AQ} 8.3 \times 6 \mathrm{BA}, 2 \times 6 \mathrm{BE} 6$. 1x6BL8. 6AL5. 6AQ5, 6CA4 and OA2. Steel case $17 \times 7+x 10 i n$.
For $200 / 250 \mathrm{v}$. A.C. mains. Brand new with full instruction manFor 200/250v. A.C. Mains. Brand new with full instruction man-
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51 n . Long play, 900 ft . $10 /-$
$5 \$ \mathrm{in}$. 1 ong play, 1.200 ft . $12 / 6$
7 in . Long play, 1,800 ft. $15 /-$

3 in. 300 ft .
Pin. © IBASI. P.V. IBASL
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 With full instructions.
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Valve line-up: two LIP88; one ECC83; one GZ34; two KLS44.
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Parforated cover with onrylug handies can be pronided if required phoo
21 /-
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12.5 K to 25 K 10 m ．
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 min．slide d．p．d．t．，8／6．

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| GET， |  |
| 89ALL．250－0－250， 60 mu， 6.3 จ． 2 a $. .17 / 6$ |  |
| 87D． $25000-250.05 \mathrm{~mA}$ ， |  |
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 EMI，Double Cone Leramic magnet 10 w ． $13 \mathrm{t} \pm$ gin． 16／．Hurn Tweeter $104 \mathrm{dl3} 2 \mathrm{Kc} / \mathrm{K}$ to $1 \mathrm{~h} \mathrm{Ke/h} 8 / C.$, STENTORLAN HP101g 10in．S to 26 ohmy 10 w． 2\％／： 8 m ．LF81：，76／－1 Crossavar Ox 8000 ， $8 /=$ EXTENSION BEPAKER CABINET．Sju．，1\％／申：

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 $1,000 \mathrm{v} .(1.001,0.002,0.005,0.01,0.02,1 / 4 ; 4.06$ ， $0.18 /-; 0.28,0.58 / 5$
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p．4－Why 2 wafer long spiadle
a p．2－way or 2 p．toway long spindia
4 p．2－way or 4 p． 4 way long upindlo 3 p．4＊way or 1 p．1z－way lony ppladie ．．$\quad$（ $/ 1$ Wastehange＂MAKTTS＂Warers Stiell－ 4 p． 3 way． 6 p． 2 way． 1 wafer switch， $8 / 6$ ； 2 wafer switoh． $12 / 8: 3$ wafer switch， $16 /=$ additlonal wafers up to $12,2 / 6$ each extre． Gaiveholdeps，EA50，6ed，B12A．CRT，1／8， B9A，gati Ceramic EFb0， $37 \mathrm{H}_{\mathrm{G}}$ ，B9A，int oct． 1／．B7G，B9A cans 1／－each，Valve plugs B7G． BaA，int．octal， $2 / \delta$ ．

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MIKE TRANSFORMERS， $5(1-1,3 / 8$ P．V．C．Covered Wire，siugle or stranded，2d．5d， Sleeving， $1, \underline{2}$ ir 4 min， $3 d$ ； 6 mm．， $5 d$ ．$y d$.

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Yoi. XL No. 696 FEBRUARY, 1965


## FELLOWSHIP OF AMATEUR RADIO

()NCE again, as we enter another brave New Year, the annual P.W. Film Show is upon us. Now very firmly established as part of the P.W. calendar, it continues to attract "full houses" year after year. There are, of course, a number of obvious reasons why this event is so popular and you can be sure that the social aspect has some bearing on its success.

Which leads us to ponder on the part that social contacts have in this hobby of ours. Amateur radio enthusiasts are naturally friendly and sociable and there is no doubt that by its very nature the hobby stimulates this frame of mind, while at the same time depending on it for success.

Other hobbies, we agree, draw their adherents together by the bond of a common interest, yet surely in amateur radio there is more to it than that? The familiar cliches about the Ham Spirit and World Friendship Through Radio may sound trite or naive to the sceptical but there undoubtedly exists an atmosphere of mutual co-operation and camaraderie in the amateur radio movement which is perhaps unrivalled.

One of the firm foundations on which this is built is the great network of local radio clubs-local RSGB groups, clubs affiliated to the RSGB and other societies, those with no particular ties, works radio clubs, school societies, etc. They all present the opportunity to mix with fellow enthusiasts in practically all walks of life, to participate in club activities such as contests, constructional programmes, discussions.

Those who stay with amateur radio (and most do) cannot fail to absorb the spirit which prevails. Those who were glad to receive a helping hand when they were learning the ropes, in turn encourage and advise yet newer recruits.

Visiting a strange town, or having moved home, one can be sure of a welcome at the local club or at the homes of local enthusiasts. A good deal of visiting takes place between amateurs of different countries, including exchange holidays. There are organisations, run by amateurs, to help the less fortunate who are bedridden or disabled. There are schemes, sometimes by groups or even by individuals, to repair radio sets for old age pensioners.

Even the "lone wolf" is never so isolated as might be expected. In the first place he has the choice of a number of lively periodicals to keep him in touch and keep the spark of interest kindled. Many have pen friends or swap tapes. Many more are members of specialised short wave societies, who issue news sheets and organise services-mostly on a non-profit basis.

Those with short wave sets can keep in touch with the whole world. And those who acquire a transmitting licence can not only hear the world but talk to it as well (when the bands are open!).

Yes, the "ham spirit" is a real thing. Whether it really contributes to world peace and fellowship in a materially effective way is perhaps open to dispute. But it certainly does more to bring nuutual tolerance and understanding than many activities on a higher level!

And we can be sure that the average amateur radio enthusiast will continue instinctively to promote goodwill in his own quiet way, not as a duty or an obligation but simply because that's the way he is.

Our next issue dated March will be published on February 4th

## New Semiconductor Light Source

NEWS AT HOME<br>AND ABROAD

SOMETHING new in light sources has recently been placed on the market by Ferranti Lid. The new device is based on radiative recombination at a $p-n$ junction, and subsequently has been termed a crystal lamp. In fact it is a "semiconductor light" which in operation acts as a forward biassed gallium phosphide diode being suitably doped to produce electroluminescent radiation of 7,000 anstroms (red).

Its physical appearance is of a cylindrical -encapsulation of transparent plastic with two wire leads emerging from one end (see accompanying illustration).

Ferranti claim many potential applications for the lamp, in industry generally but particularly in instrument and indicator applications where its small size -typically $0.03 \mathrm{in} . \times 0.04 \mathrm{in}$. and low operating currents, are particularly advantageous. An obvious environment is transistorised equipment in which the voltage and current levels are suited to its use as, for example. a simple on/otf indicating device.


This illustration makes clear the construction of Ferranti's new crystal lamp-a semiconductor light source. The average brightness of this unique lamp is $10-40 \mathrm{ft}$. lamberts and it remains visible under an illumination of 20 ft . candles.

## NEW NATO AERIAL IS EUROPE'S LARGEST

THE largest aerial system in Europe has recently been completed at Anthorn, near Carlisle. The aerial has been erected for the North Atlantic Treaty Organisation's Anthorn v.l.f. radio station and consists of 13 masts, arranged as a " six-pointed star" with a centre point, stupporting a network of diamond-shaped panels forming the aerial (see illustration below).

The main contractor for the station was Continental Electronics Systems Inc. of Dallas. Texas, working to the requirements of the British Post Office, with British Insulated Callender's Construction Company Limited undertaking the design, supply and erection of the masts.

The outer ring of six masts-each 618 ft . tall-is 2.148 ft . in radius, with the inner ring of 678 ft . tall masts having a radius of $1,277 \mathrm{ft}$. The central mast from which the aerial panels radiate, has a height of 748 ft .

A further impression of the immense size of the system may be gained by considering the specifications for the station earthing system. This is virtually a copper mat covering the whole of the site, 9 to 12 in . below ground. Lines of 8 s.w.g. soft copper wire radiate from the centre mast with $2^{\circ}$ of arc between each. requiring in all, a total of 75 miles of copper wire.

The construction of the aerial


Looking down on the Anthorn v.l.f. station from one of the outer masts, showing ore of the massive insulators in the foreground.
panels called for some 20 miles of steel-cored aluminitum conductor, which were cut and prestressed on the site before their assembly. At each corner of each panel, 40 ft . long irtsulators support the aerial. These insulators are fitted with two corona rings, one 15 ft . in diameter, the other 12 ft . in diameter.

## The Radio Show Goes International

FOR the first time ever, this year's Radio Show at Earls Court,
London, will he international in scope. Ever since the days of Radiolympia, it has always been a national show with a British-made-only policy. Now it is hoped that exhibitors from abroad will help boost attendances.
The promotion and organisation of the annual exhibition has been taken over this year hy Industrial and Trade Fairs limited from the previous organisers, the Britich Radio Equipment Manufacturers' Association.

The dates of the 1965 Radio Shows have already been announced as 25th August to 4 th September.

## 100kW Transmitters for Malaysia

'THE Gates Radio Company. a subsidiary of the Harris-lntertype Corporation of the U.S.A., has recently completed a Government of Malaysia contract for three high powered 100 kW short wave broadcast transmitters. The transmitters will be the most powerful in Malaysia. and will be used by Radio Malaysia for longrange international broadcasting to inaugurate its new "Foreign Broadcast Service" to neighbouring South-East Asian countries.

## RENOVATIONS AT RUGBY

THE 16kc/s world-wide trans mitter at Rugby (callsign G.B.R.) is to be renovated.

This was announced recently by the G.P.O. and now another annourvement from the English Electric Valve Company Limited tras confirmed that its output stage will be re-equipped with EEV BY1144 triodes. This modification means the replacement of the 54 water-coolled triodes. type CV1600 previously installed in three banks of 18. where two banks provided the 210 kW output and the third remaired on standby: All these will be replaced by only three of the English Electric valves, these being vapour-cooled power triode types. two of which will provide double the original output and leave the third on standby.

These transmitters are of a new air-cooled design, employing high level plate modulation and utilising conservatively rated components especially chosen for reliable service in areas of extreme temperature and humidity: conditions such as may be expected in Malaysia,

## HEREFORD'S NEW RELAY STATION

THE sound transmitters at the BBC's v.h.f. sound and television relay station at Hereford were brought into service during November last year. Transmissions of the Midland Home Service is on $94.1 \mathrm{Mc} / \mathrm{s}$. the Light Programme on $89.7 \mathrm{Mc} / \mathrm{s}$. and the Third Programme/Network Three on $91.9 \mathrm{Mc} / \mathrm{s}$. Horizontal polarisation is used.
The new station will improve v.h.f./f.m, reception for the 40,000 . people in the City of Hereford and its environs.


# BUILDING A 

 TRANSISTOR SOLO organAFTER many experiments with various types of audio oscillator, the author decided to build a musical instrument, using the well-known multivibrator circuit. This will be familiar to many readers, as it is frequently used for a signal probe in radio fault finding.

In the instrument about to be described, instead of the usual thin reedy tone, we have a really musical output which can be fed into any amplifier and will provide tones resembling the oboe, horn and cello.

Furthermore the tones can be varied with slow, medium and French or continental vibrato (the latter resembling a French accordion).

Percussion is also provided with (1) fast decay; (2) slow decay; thus simulating plucked strings of banjo, guitar, etc.

## Note Generator Circuit

The circuit (Fig. 1b) will now be explained. This is divided into two sections, as recommended in actual construction. Section 2 is the musical note generator-a straightforward multivibrator except that instead of one, we have a series of resistances in the base of Tr4.

R11 is a limiting resistor to protect this transistor. VR2 is a variable resistor used for setting up the tuning to correct pitch. For convenience it is brought out handy to the keyboard.

Playing the highest note on the keyboard connects VR2 to the negative line and sounds the highest note. The remainder of the resistors in the series are tapped off to the key contacts; and when the bottom note on the keyboard is played, the whole series will be in circuit and sound the lowest note.

The range of the author's instrument, as described here, is 29 notes from $C$ in the bass to $E$ in the treble (see Fig. 2).

It will be appreciated from the foregoing why chords cannot be played, and if more than one key is played together, only the uppermost note will sound. This is a very practical advantage in a monophonic organ such as this.

The series tuning has some disadvantage and it
is not desirable to extend the range for any single oscillator much beyond the two octaves.

## Vibrato Circuit

The :base of $\operatorname{Tr} 3$ receives the low frequency impulses, generated by the oscillator (Section. 1, Fig. 1), which provides vibrato. The $1 \mathrm{M} \Omega$ variable resistor VRl is a preset by which adjustment can be made to obtain a slow vibrato.

One turn of the vibrato switch places R3 in parallel with VRI and we obtain a medium speed: one more turn of the switch and we have a fast speed. It is possible that these two resistors may need: some slight alteration of values owing to difference in tolerances; however, it would be easy to, experiment here as they are soldered directly on to the switch. Obviously slow speed must be correct before alterations are made.

The vibrato switch Sl (Fig. 1c) is mounted under the keyboard for easy adjustment whilst playing. Another switch S2, cuts out vibrato when not in use.

## Buffer Stage Circuit

The output of the note generator is taken from base of Tr 3 via the $0.01 \mu \mathrm{~F}$ mica coupling capacitor C6. This gives a more useful waveform for our purpose than if taken from the collector of Tr4. A mica capacitor is preferred here for stability.

This output is fed into a buffer stage (Fig. 3a). otherwise changing over of tone circuits would affect the pitch.

## Tone Circuit

This - is shown in Fig. 3a and follows usual practice except that instead of using three separate stops it was decided to use another 3 -way switch (S3).

This way the tones cannot be added as in usual organ practice, but experiments proved that there was nothing to be gained by so doing.

## Pre-amplifier and Percussion

The final stage is a pre-amplifier which, with the aid of one extra contact per key, has the
additional function of providing percussion. Thes is a trick circuit and with the key action devised by the author works very effectively.

In normal plasing L.-M1 (Fig. Sa) si shorted (midposition S4A-see Fig. 3a). The percussion key contacts. although still working. are now inoperative. Only the key primary contacts (Fig. Ib) are now effective, being normally open and contacting the -3V (No. 1) negative strip when keys are pressed down in playing.

For percussion. short circuit on $I-M$ is removed and replaced by a $1-2!\mathrm{F}$ capacitor. Current is now diverted via the secondary key contacts (Fig. 3a) which are normally closed. at the same time

It would seem a good idea to switch in a forther stage of amplification when using percussion. However. as this would lead to further complications, it was decided to do the reverse and cut down volume on normal playing-after all we shall have plenty of amplification to spare in the main amplifier which follows.

When changing over from percussion to normal, switch S 4 B will close and bring R30 in parallel with the lower part of the expression control. This will reduce the volume according to the value of R30 which. in the writer's case, is 20 kS . This works quite well, once adjusted to a level volume.

VR3 (Fig. 3a) is the expression (volume) control,


Fig. Ib, The note generator stage (Section 2).


Fig. Ic: Details of the SI suitch connections.

Fig. 2: The tonic sol-fo range of the organ.

$\therefore$ in the writer's case controlled by a foot pedal to be described later from which a two yard length of screened cable can be plugged into a socket fitted to the organ case.

## Connecting to Main Amplifler

Another screened socket of a different typeto avoid accidental interchange-will receive the screened lead from the constructor's amplitier.

This can be a transistor amplifier, in which case the coupling capacitor CI7 can be increased in value to $10 \mu \mathrm{~F}$.

It is not advisable to use an amplifier with live chassis, but if one is used a high voltage $0.005-$ $0.01 \mu \mathrm{~F}$ capacitor should be inserted between the organ and chassis. This will probably give rise to some hum, which however slight, will mar the tone of the organ.

Should there be a volume control on the amplifier, this should be set at full volume when
the pedal is fully depressed and then left at that setting. A tone control, if fitted, can be adjusted to one's liking.

## Design of the Case

The writer's instrument is built around a discarded accordion keyboard of 29 notes. The note generator is designed to sound the notes from C in the bass to F in the treble (Fig. 2).

The case measures 1 ft . by 1 ft . 2 in . and $4 \frac{1}{2} \mathrm{in}$. deep. A leather strap handle is fitted at one end. The amplifier which contains an 8 in . speaker is of similar size also with a handle, so that the outfit is quite portable.

All organ controls are brought out to the front, just below the keys; and each end of the case has a length of $\frac{3}{4} \mathrm{in}$. $x$ fin. aluminium angle, so fitted that two flanges can slide upon two runners fitted under the treble of a piano.

The expression control being pedal' operated leaves the left hand free for accompanying on the piano. With the cable connected to the amplifier, the outfit is ready for use.

Alternatively the instrument could be designed to play as an accordion. In this case, the contrals (including expression) could be arranged for the left hand and then only needing one screened cable to the amplifier.


Fig. 3a: The buffer stage, tone and preamplifier and percussion sections of the circuit.


Also it could be designed as a solo (upper) manual. to use with a single manual electronic or reed organ.

## Power Supplies

Battery No. 1 (3V) supplies power for vibrato and note generator. Battery No. 2 (3V) supplies the buffer stage. Percussion and pre-amplifier stage is powered by two similar 3 V batteries connected in series to give 6 V .

All positives are common and taken to a twopole on /off switch $S_{5}$ and wired as shown in Fig. 4. Although they are shown as being together, it is preferred to secure each battery with a clip near each panel.

A two-pole switch is necessary owing to the 3 V difference in potential between the batteries. Type No. 1839 3V batteries have been used throughout, and as they last at least a year, all connections are soldered.

It is recommended that the two cells in each battery are pushed out of the case and the negative of one cell connected to the positive of the other with a short length of thin stranded flex and then reinserted into the case.

Remember that the cell case is negative. especially before finally connecting to their respective panel tags, otherwise damage will result to the transistors.

## Wiring Note Generator (Fig. 5)

It is advisable to start by wiring up the note generator, for which a $\pi_{16}^{16} \mathrm{in}$. paxolin panel about 3 in . $\times 4 \frac{1}{3} \mathrm{in}$. is required.

The components are mounted on this panel, the wire ends being cleaned and bent over at right angles and pushed through holes in the panel for wiring up on the underside.

Thin insulated connecting wire should be used and where a junction has to be made, if a piece of systoflex of a sliding fit has been previously slipped over the connecting wire covering, this can be pushed down over the junction after it has been soldered. Leave transistor leads full length, even if they have to be long routed.

All wires should be cleaned and tinned before soldering. A heat shunt should be applied to the wire between the iron and transistor, as well as wires to small electrolytics.

A common dodge is to use a pair of tapered flat-nose pliers and stretch a rubber elastic-band across the handle, thus leaving both hands free for soldering.
" Red spot" transistors have been used throughout with the exception of this note generator, lor which we desire good stability in order that musical pitch will remain reasonably constant with varying temperature. After experiments it was decided to use the Mazda XB102 transistors which were found to be very stable.

Before drilling the panels it is a good idea to lay out the components on a piece of cardboard of same size as the panel, piercing holes where required and trying the layout, which should include space for five soldering tags.

When satisfactory, the card can be used to mark out the holes on the paxolin for subsequent drilling. Five soldering tags should be fitted at one side. convenient for external connecting and with a sharp pointed awl, the following legend

should be scratched on the paxolin adjacent to the appropriate tag: + VIB.IN. O.P. KEYS-No. 1 3V O.P.; being of course for output from that panel.

Care should be taken that the transistors are wired correctly, as layout is reversed upon turning the panel over for wiring: and it is therefore suggested that the letters C B and E are scratched on the underside of the panel against each hole through which these wires will pass from each transistor.

Assuming that all wiring has been checked and found correct. it will now be possible to try out the uscillator by connecting tags + and O.P. to high resistance headphones.

A variable resistor of about $5 k \Omega$ should be connected across tags marked keys and - No. 13 V . Then connect the positive of a 3 V battery to the + tag and if a lead is taken from the battery negative. this when touching the - tag should produce a sound which can be varied in pitch by rotating the variable resistor.


Fig. 5: Wiring and layout of the note generator.


Fig. 6: Vibrator wiring details.

## Wiring the Vibrato Oscillator (Fig. 6)

This panel can be of same size and wired up in a similar manner. The $1 \mathrm{M} \mathrm{\Omega}$ potentiometer VRI is a skeleton type preset and is mounted on the upper side of the panel.

Four soldering tags should be conveniently positioned and bear the legend: + O.P.(Y) O.P.(X) $-(Z)$.

The wiring of the three-way switch is clearly shown in Fig. lc. As drawn it is in the slow
clicks should now be heard, if not the $1 \mathrm{M} \Omega$ potentiometer may need adjustment.

If there are still no clicks. set the potentiometer about halfway and place a fixed resistance of about $470 \mathrm{k} \Omega$ in parallel with same by spanning $-Z$ and $Y$, again slowly rotating the track arm.

If no clicks are heard, it is possible that one or both transistors are unsuitable and substitution should be tried. When working properly it should be possible to obtain four to five beats per second, without the fixed resistor in parallel.


Fig. 7: Layout of components and wiring connections of the buffer stage, tone and preamplifier panel.
position, next is medium and finally fast (French vibrato). The purpose of $\mathrm{R} 74.7 \mathrm{M} \Omega$ ) is to increase the amplitude or depth of slow and medium vibrato.
The layout of the wiring in these panels is not critical, but panel inter-connections should be as short as possible and miniature microphone screened cable has been used as much as anything for its tidiness.

This also applies to connecting up switches to papels, including vibrato negative and on and off switch. This latter is in the negative lead from battery No. 1, a lead from the switch going to -(Z) on vibrato panel.
This panel can now be tested (without the threeway switch) in a similar manner to the note generator, but in this case we connect a $0.1 \mu \mathrm{~F}$ capacitor to tag O.P.(X) the other side of the capacitor to one side of headphones, and the other side of headphones to + tag.

A 3 V positive battery is now connected and a lead from the negative of the battery, preferably with a crocodile clip connected to -(Z). Slow

Buffer Stage, Tone Circuit and Pre-amp (Fig. 7)
This is our last panel (Fig. 7) and for a suggested layout see Fig. 7. The soldering tage are arranged to suit the writer's particular layout and may be found suitable by other constructors.
The buffer stage must have a separate battery as shown. $\Gamma 1$ is a driver transformer as used in small transistor radios with push-pull output. A capacitor is wired across one winding and the centre tap ignored. T2 is a small transistor type interstage transformer and one winding is left open.

There is scope for experimenting here, with whatever types one has to hand. Switches S3 and S4 tre brought out to the front of the instrument under the keyboard.

R21, R22 and R23 could of course have been wired as a single resistor of $47 \mathrm{k} \Omega$ between switch arm and C12. but this layout was adopted as these three resistors also serve to level off the volume of each separate tone output as may become necessary with other transformers.
next month - the keyboard and cabinet

## PART ONE



BY F. L. THURSTON

IT is the object of this short series to deal with the " non-electronic" side of the hobby; what can be loosly termed the "Ironmongery" side of it-chassis construction, cabinet making and finishing.

The series is aimed primarily at the novice, but more experienced readers may find much of interest, particularly in such matters as glass fibre cabinet construction, chemical finishing of metals, ctc.

## Chassis Construction

The most elementary type of chassis is the "breadboard". consisting simply of a slab of wood on which components and controls can be easily mounted when making temporary hook-ups or testing experimental layouts.

Components such as transformers, chokes, tag strips, tetminal blocks, etc., ate fixed with ordinary wood screws. Valve bases and controls may be mounted on additional strips of wood or metal at right angles to the breadboard. "Bus-bars" may be used for h.t. and negative return lines.

An improvement on the conventional breadboard is to cover the entire upper surface with a single sheet of aluminium or copper foil, glued to the wooden surface with a suitable compound. Such a board provides electrical continuity, for earthing, etc.

These systems should be used only on temporary lash-ups, as the boards will be subject to heating and age and will bend or crack after prolonged use, resulting in unstable operation.

For permanent set-ups, excepting for transistor equipment, an aluminium chassis is preferable. A
number of manufacturers produce ready-made chassis of excellent quality and reasonable price. Nevertheless, many readers will prefer to make their own, and the following notes may be of some use.

A very simple method of construction, which involves no bending of metal at all, is shown in Fig. 1. Sheets of aluminium are cut to form the top and sides of the chassis and are bolted together with pieces of angle aluminium along the edges of the sheets. Additional lengths of angle may be used where extra rigidity is needed.

The angle should be between $\frac{1}{2}-\frac{3}{4} \mathrm{in}$. in width for all but the exceptionally large or small chassis. and the aluminium used for the chassis should be 20 s.w.g. up to about $2 \frac{1}{2} \times 3$ in., 18 s .w.g. up to about $9 \times 7 \mathrm{in}$. and 16 s.w.g. for sizes larger than this.

The most generally used chassis is the " single sheet "type. and Figs. 2a 2b show two alternative ways to produce the same size chassis. Fig. 2a is probably the more ditficult but more effective of the two methods. As well as bending to form the sides it is also necessary to add the four "lips ", marked A, B, C and D.

When the chassis is finally formed. these lips must be on the inside of the two shorter side pieces and should be drilled through and bolted or riveted to make the chassis a rigid structure. In Fig. 2b these lips are not used, but separate pieces of angle or other strengthening fixtures take their place to give almost equal rigidity.

## Bending

The actual method of bending the metal will depend on the facilities available to the construc-


Fig. 1: An underside view of a chassis assembled from flat aluminium plates bolted to aluminium angle. This form of chossis construction involves no bending of metal at all.


Fig. 2a: With this single-piece chassis all holes are drilled AFTER bending is completed, and the line of bend of lips is set back from main bend-see text.
tor. For those that can afford it, special sheet metal folding machines are available and are a good investment. Those for amateur use range in price from about $£ 610 \mathrm{~s}$. to $£ 12$.

For the average constructor. a good vice may be used as a bending tool. Smooth faced jaws should be used to avoid marking the aluminium unduly when bending it.

First cut the aluminium sheet to size and score with a scribing tool along the proposed lines of bend, on the underside of the sheet. The lines should not be scribed too deeply, or it may be found that when the bending is nearly completed fatigue fractures will occur along them.

If, say, the vice is of the 4in. type the bending procedure for Fig. 2a will be as follows: insert lip A into the vice and ensure that the bending line is flush with the top of the vice jaws along the whole of its length, and tighten the jaws.

Note that this bending line falls behind the bending line for the short side piece by the thickness of the basic material.

Apply pressure and bend the metal over, making sure that the scribed line is on the underside, otherwise the lip will tear away or fracture along the line. A soft mallet can be used to help the bending operation along.

A final sharp corner is obtained by placing a piece of scrap aluminium over the bend area and hammering down on it until the desired result is obtained. Repeat the operation with the remaining three lips.

Next, bend the 5 in . sides, and finally the 10 in .


Fig, 2b: With this form of construction, "lips" are re placed by aluminium angle. After bending, the sides of the chassis are secured with the angle pieces.


Fig. 3: A simple device for exterding the effective length.of vice jaws.
sides. When the side to be bent is longer thian the vice jaws it should be bent in a number of stages, starting with a small angle of bend at the left and then working along the length of the side until the same angle of bend is, obtained throughout, returning to the left again with a greater angle of bend, repeating the procedure until the final sharp angle is obtained.

The effective length of the vice jaws can be extended with the simple gadget shown in Fig. 3. Made from two lengths of angle iron or steel and held parallel by bolts passing through the two guide holes, the gadget can be made even more useful by filing the top faces flush with each other and case hardening the whole assembly.

The device can then be used when sawing or filing a straight edge along aluminium, etc., as well as for marking out and bending.

Fig. 4: Two types of hole-cutting tools for chassis work. Holes of several inches in diameter down to holes less than one inch in diameter can be cut easily in aluminium sheet with both tools.


## Holes in Sheet Aluminium

A number of manufacturers produce punches that cut holes of fixed size, ranging from about $\frac{3}{8} \mathrm{in}$. to $2 \frac{1}{2} \mathrm{in}$. diameter and ranging in price from a few shillings to a couple of pounds each. Usually of the "screw down" type, with a punch in two parts drawn together by a bolt, a pilot hole must be drilled through the aluminium before the punches can be used.
At least one manufacturer produces a doubleended punch so that one gets, in effect, two punches for the price of one. Punches are available for cutting square as well as round holes.
The advantages of these punches are the ease with which holes can be cut with them and the fact that almost no "cleaning up" is needed with a file afterwards. The major disadvantage is their cost.
Another tool that is available for cutting round holes is shown in Fig. 4. Consisting of a central drill and a bar set at right angles to it, with an adjustable cutting bit on the bar, the device enables


Fig. 5 (left): A simple method of cutting rectangular holes.
Fig. 6 (right): A case-hardened template for cutting Itin. valve base holes.

When cutting rectangular holes in sheet metal the method that most text books recommend is shown in Fig. 5. The size of the required hole is carefully marked out with a scribing tool and a second set of scribe lines are drawn about $\frac{1}{16} \mathrm{in}$. inside of the first set.

A $\frac{1}{8} \mathrm{in}$. drill is then used to drill a complete series of holes along the entire length of the inner lines until the entire strip is cut away and the centre section can be pushed out. The jagged edges of the hole are then carefully filed down to the outer set of scribe lines and the hole is finishepd off. Note that if a $4 \times 4 \mathrm{in}$. hole is cut by this method a total of 128 small holes must be drilled!
A better method is to first drill a $\frac{1}{2}$. hole in each corner after marking out the outer scribe lines, and then simply cut out the hole using a fret saw with a metal cutting blade, or a "Keyhole" or pad saw. The whole centre section can thent be removed within a couple of minutes!

The metal cutting blades can be purchased very cheaply from most tool shops and are surprisingly strong in spite of the fact that they are less than $\frac{1}{12}$ in. deep. They could, in fact, be used to cut a hole only tin. square if required!
The least expensive method of all for cutting large round holes is to make up a template in about tin. sheet steel which is then case hardened. In use the template is bolted down to the chassis and the hole cut out with the fret saw. Fig. 6 shows such a template; made for cutting $1 \frac{1}{8}$ in, valve base holes.

Manufacturers of valve bases do not seem to have standardised the radius from the centre hole of the base securing bolts, so provision has been made for three sets of bolts at different radii. A slot could also be used here.
holes ranging in diameter from less than an inch to several inches to be cut with a single tool.

Some of these adjustable cutters are provided with a bar extending on both sides of the drill and have a bit on each extension, resulting in a smoother cutting operation (Fig. 4). Both types are available from most radio and tool stores.

With these tools the holes need to be cleaned up with a file after cutting, but they are an excellent investment and are the only tools that will cut large diameter holes reasonably well.

When drilling holes in aluminium the drill will tend to wander and unless care is taken the final hole will be off-centre. It is therefore advisable to start off with a small pilot drilling to reduce this wander as much as possible, and to cut the hole in a series of progressively larger drillings rather than in one single operation. For a similar reason, power drills must not be used unless they are mounted in a stand.

When cutting holes for valve bases, etc., and punches are not to hand, a carpenter's bit makes an excellent alternative. This tool, like the adjustable hole cutter, is best used mounted in a carpenter's brace.

## Printed Circult Techniques

Printed circuit boards consist of a sheet of copper foil bonded to a sheet of Paxolin or similar insulating base, surplus foil being removed by etching. Un-etched boards are available from- the larger radio stores, but are fairly expensive.
The experimentally minded constructor can make his own boards by bonding a sheet of copper foil to a suitable base, using a waterproof adhesive such as Durofix, Araldite, etc.
The base should preferably be Paxolin or a similar type of insulation, but less expensive materials, such as 3 -ply wood, can be used, providing steps are taken to ensure that they are made non-porous. This can be done with plywood by first thoroughly drying it in an oven at a fairly low temperature over a long period of time and then soaking the assembly in a good quality varnish. The base must have very good insulating properties.
An alternative to copper foil is aluminium foil with an electro-plated copper deposit on it. The plating can be carried out with a minimum of facilities. Bond a sheet of aluminium cooking foil to the base; and, after it has throughly dried, clean

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the surface lightly with a piece of fine emery paper, removing all traces of oil and oxidization.

Mix a solution of 2 oz. copper sulphate to a pint of water in a shallow dish of glass, plastic or china. Place the board in the solution. face upwards. and connect a lead by crocodile clip between the foil and the negative terminal of a low voltage battery (less than 6 V ). Connect the positive terminal to a loop of heavy copper wire suspended over the surface of the foil but not in contact with it, the loop to be submerged in the solution.

As soon as the final connection is made, bubbles will rise from the copper, indicating that electrolysis is taking place. After 5 or 10 minutes the foil should be covered with a fine coating of copper and can be removed from the solution and washed under a tap.

The plating should not rub off under normal finger pressure. If it does it will be because either the surface was not cleaned fully at the beginning of the operation or too high a voltage has been used, resulting in excessive current flow and soft plating. After plating, the board is left to dry at room temperature and it is then ready for use.

The printed circuit board. home-made or purchased, is now ready to be prepared for the etching process. When the layout of the board has been decided, it should be carefully marked out with the pattern and the position of the holes for the components determined. A drill, about $\frac{1}{82}$ in. diameter, is then used to drill gently through the board from the copper side.

The area of the board not to be eaten away in the etching process is now painted over with an acid resistant paint, such as Valspar, or with nail varnish, which is very quick drying. Make sure that the paint covers the copper exposed by the hole-drilling process.

A solution of about 30z. ferric chloride to one pint of water, is now made up or purchased from the chemist, and placed in a shallow dish and the painted hoard gently immersed in the liquid, face upwards. The untreated metal will slowly dissolve, helped along by rocking of the dish.

When the entire untreated surface is seen to be eaten away the board is removed and washed under the tap to remove all traces of the solution. Finally, the paint is removed with a suitable spirit or paint remover and the hoard is ready for use.

The above etching process may take as long as an hour or so, depending on the strength of the solution, which must not be allowed to come into contact with the eyes or clothes.

When soldering the boards. the iron must not he held against the copper foil for ton long, or the foil might hecome detached from the hase.

When the circuit has been finally built up and tested it is often a good plan to dip the entire assembly in a good quality insulating paint or shellac, to protect the circuit from dampness and dust damage.

## Encapsulation

A method of construction often used on military equipment is to loosly assemble a circuit in three dimensions using the connecting wires to support it, and then encapsulate the entire assembly in an
epoxy resin. so that the finished product is a solid block of resin enclosing the circuit.
Such a circuit is incredibly strong, very compact, can be submerged in water with no ill effect, and is impervious to most acids. The only snag is that servicing is difficult!

If a piece of gear of this type does break down. it is necessary to chip away the resin until the faulty component is exposed, cut it away and replace it with a good part and then re-encapsulate the unit.

The procedure for making home-built units of this kind is as follows. Decide on the dimensional layout, wire the unit up and test it. If valves are used. they must he mounted external to the block, with the top of their valve holders flush with the top of the resin. Transistors may be fully submerged. Coat the whole assembly with varnish: this will facilitate salvaging of components during repairs.

Prepare a mould of tin, cardboard, or similar material to accommodate the wired-un unit. with a little space to spare all round. and line the inner surface with Vaseline, to stop the resin sticking to the mould.

Prepare the resin as per manufacturer's instructions, Bondaglass" Clear Casting Resin", costing 10 s .6 d . for 10 oz ., is recommended, available from Bondaglass Lid., of 53-55 South End, Croydon. Surrey), and pour into the mould to a depth of between $\frac{1}{8}$ and $\frac{1}{4}$., and allow to set.

When this layer has dried. lower the painted circuit on to it, keeping it central in the mould. The paint should still be wet. Pour in more resin. until the circuit is covered, and warm slightly to remove any air bubbles. The resin is traislucent and these will be visible. Place the unit aside to dry. This should take no more than 12 hours. Finally, renove the block from the mould and check that the circuit is still functioning correctly.

The main snag with this type of encapsulation is the expense. An alternative is to use Plasticine in place of resin as the encapsulating medium. The loss in strength can be made up by leaving the block in its mould. Such a technique retains the advantages of compactness. acid resistance. water proofing and ability to withstand hard knocks, withnut the disadvantage of being very difficult to service.

An alternative method of construction to the printed circuit that has many of its advantages may be used in conjuction with the Plasticine encapsulation technique, the resulting assembly being very chean to make. Instead of using a piece of printed circuit to hold the components in place, a piece of Paxolin or similar material is used.

Holes are drilled to take the components in the normal way. and the components are pushed through the holes from one side of the board and ordinary connecting wires used to wire them up on the other side.

The components are thus held firmly in one plane. hut are still subiect to movement in the vertical. Ry now encapsulating in Plasticine, all movement of the components under natural forces is finally arrested and the unit is complete.

Part Two Next Month


All times are in G.M.T.
The Broadcast Bands-by John Guttridge

CONDITIONS in the short wavebands are now at their lowest ebb, with even the $11 \mathrm{Mc} / \mathrm{s}$ band offering very limited openings after about 2100 . Things should start improving, though, as winter recedes.

Station observations come from three listeners this month. Firstly up in the Shetlands reports good reception of Radio Peking's transmission to Australia from 0800-1030 on 9.457 .

In Elgin, Paul Harris has been having some success with the 2100-2150 transmission of Radio Nederland on $6.085 / 9.590$. He has also observed Radio South Africa in the evenings on 9,525/7,270. The former frequency, he says, is good with close down at 2115 .
E. H. Conduit of Wolverhampton sends the schedule of English transmissions from Radio Berlin International (the East German state radio) to Europe. These are on the air from 1700-1800 on 6,080/6,115/7.132/7,300/9.730. At 2015-2045 these same frequencies are used with the addition of 1,510 medium wave. The last transmission of the evening is at $2200-2230$ on $6,080 / 6,115 / 7,300$ and 1,430 medium wave.

He also sends the schedule of United Nations transmissions to Europe. These are aired over Voice of America transmitters from 1800-1900 on Fridays and consist of ten minute programmes in Arabic, English, French, German, Greek and Turkish. Between $1800-1830$ the programme is carried on 21.485 (Bethany), 15.250/17.800 (Greenville) and 9,710 (Tangier). From 1830-1900 frequencies used are 15.315 (Bethany), 15,150/11,780 (Bound Brook). 9.610 (Tangier).

African stations have been giving the best evening reception on 19 metres, Voice of America, Monrovia, being audible until after 2200 with English programmes. A surprise station in this band was Trans World Radio, Bonaire, (Netherlands Antilles) which was heard testing. with a wẹak signal and asking for reports at 1915 on 15,300.

One of the few signals to reach readability 3 in the 25 metre band in the late evening was Radio Abidjan, Ivory Coast. This was heard in London in French with Music on 11.820 at 2245.

If you haven't logged a Venezuelan yet try 4.970. If c.w. interference is not too bad you should be able to hear Rudio Rumbos. Identification is given at 2230 . The address for reports. which are verified, is Apt. 2618. Caracas. Venezuela.

The Kuwait Broadcasting Scrvice. P.O. Box 397, Kuwait. currently uses two short wave outlets. These are a 50 kW transmitter on 9,520 and a 10 kW
transmitter on 4,967.5. An English programme is carried on the latter frequency from 1700-1900. An attractive QSL-folder is sent for correct reports by this station.

Extensive changes have been made by Radio Moscow in its European English winter schedule valid until Junc 4th, the only unchanged transmission being that from 1200-1230. Details of the other transmissions are $0700-0730$ on $9,590 / 7,280 / 7.240 / 5,980 ; 1900-1930$ on $7,330 /$ $7.280 / 6.100 / 5.980 / 1.320 ; 2000-2030$ on $7,330 /$ $7.280 / 6.100 / 6.200 / 1,380 ; 2100-2200$ on $7.330 /$ $7,280 / 7,260 / 6,050 / 1,490 ; 2200-2230$ as 2100 plus $5.960 / 1.380 / 1.320$.

Radio Japan. Tokyo, Japan, now uses 7,195 in its 1500-1630. 1645-1845 and 1900-2000 transmissions to South Asia, the Middle East, and Africa respectively. The other frequency used is 9.525 with the addition of 9.705 for the 1500 transmission. General service transmissions between 1200 and 1930 (the first half of each hour) are now on three 31 metre band outlets 9,505/ $9.605 / 9,740$.

Full verification details are given on the QSL issued by Radio Voice of the Gospel, ETLF, P.O. Box 654. Addis Ababa, Ethiopia. Unfortunately it neglected to send a progranme shedule. This station transmits 12-13 hours of shortwave broadcasts in 13 languages every day over two 100 kW transmitters.

The date is the only verification detail included by all India Radio, Post Box 500, New Delhi, India, on its otherwise colourful QSL Card. Several changes have been made in this station's winter schedule valid until March 6. The English transmission to Australia from 1000-1100 is now on $11.710 / 15.165$ with $9,655 / 11,700 / 15,105 / 17.855$ being used at the same time to Asia. The East African English transmission from 1840-1930 is now on $7,180 / 9.680 / 11,815 / 11.940$. Although the frequencies-6,130/7.235/9,915—of the European English transmission at 1945-2045 remain unchanged, the relay to West Africa at the same time is now on $7,125 / 9.690 / 11,740$.

Transmissions to the Caribbean and South America in Portugese from 2331-2400 and Spanish from 0000-0045 from the Canadian Broadcasting Corporation, P.O. Box 6000. Montreal. Canada, are now going out on 11.760 / $9.625 / 5,990$. This winter this station not scheduled to use a frequency lower than 9.630 to Europe. although it used the 49 metre band last year. Full verification is given by this station to correct reports.

Monthly programme schedules are not to be distributed by Radio Nederland, P.O. Box 222. Hilversum, Holland, in future. Instead schedules will he issued four times a year to coincide with the changes of season and frequencies. During January and February programmes remain basically unchanged with Dx Juke Box on Thursdays. Frequencies are as December.

The winter schedule of the Voice of America gives the following frequency usage for English to Europe: $3.9800300-0730,1400-2345 ; 5.9950300$
-0730. 1630-2245; 6,040 0500-0730; 6,080 0300 -0730: 7.200 0300-0730: 7.205 1500-1800, 1830 —2245: $9.5450500-0730 ; 9,6701915-2215 ; 9,740$ $0500-0730$. $1830-2245 ; \quad 11.790 \quad 0500-0730:$ $11.8251900-2215 ; \quad 15,205 \quad 1400-2215 ; 15,290$ 1400-1630; 17.780 1400-1800. The medium wave outlet on 1,196 is used from $0400-0430,0500-$ 0730. 1700-1830. 2200-2345.

Finally, a collective "thank you" to many readers who have sent along news and reports. Sorry we cannot mention you all!

## The Amateur Bands-by David Gibson G3JDG

FIRST this month a big "Thank you" to all those SWLs who sent in reports on the various bands.

## $1.8 \mathrm{Mc} / *$

Top band seems to be rather abandoned by most reporters and no one appeared to listen out very much. I had a couple of very hectic one-hour sessions on November 14-15th in the M.C.C. 1.8 Me/s contest. Many $G$ stations were worked and the best "DX" was GWGGW at Blackwond in Monmouthshire. A ten-minute listening period. Sunday, November 29th. between 1800 and 1810 , logged G2PT, G3SYX, G3TLE, G3OCA, PAOPN. GC3ECC. DJ3JZ.

Master Darrell Earnsháw (Preston) scems to have things pretty well organised in that his sister does a turn at the receiver. They offer the following log for $1 \cdot 8$ : G3EXU, G3FIF, G3DMO. G3SSU. G3RFN. G3KE, G3NVN, G3GGS. G3TNN, G3KKU, G3WLQ, G3GEU, G3SWM. G3HFD. G3TFN, G3ERY, G3BES, G3JMA. G3NSW/M, G6ES/M, the stroke $M$ signifying that the station is mobile and operating from a vehicle. All these stations on phone.

### 3.5Mc's

This band also seems passed over by most listeners and only one letter offers any log at all. T. Cridland, of Canvey Island. uses a Codar CR45 (TRF) with an 18 ft length of wire mounted vertically up the side of the house, heard these on November 14th hetween 0932-1016: G3PI.R. G3HRD. G3¢J. G2WJ. G3MY. GこBC4. G3P7X. Novạmber 15th (1220-1342): G3PIX. G3RAO. G3POC. G3GFJ. G3PMC. G3PFT. G2AFR. G3RBF, G3MUW, G3SNI. G3TIZ.

## 7Me/s

Forty metres also is sadly hypassed hy all and sundry and no one mentioned it this month. It must be confessed that a quick listen most evenines would deter all but the very brave. This is a wame hecause after a while odd call signs start to appear and a little perseverance can produce quite an interesting log. especially early mornine. Ten minutes on Nowemiber 29th (1800-1810) leaged LZ?ANB. Gi3GIX. YO8KAF.. ITIKBD. YO5KAI, PAOIV, ITR2AO, IB5KCE and C'R6AT.

## 14Mc's

And so to the h.f. bands where 20 and 19 m are firm favourites-not surprising considering the volume of traffic, particularly on 20. From 1738-

1746 (yes. just eight minutes) W4BV, W1BFA, W2KXV. W2LSW, W2AZS, W3BYX, W4KFC, W2PCJ. IIZCN. K3JC'T, GB2SM, K8IKB. KIBBV were heard and these were only the loud ones.

Immediately after (at 1748) the bandswitch was set to $21 \mathrm{Mc} / \mathrm{s}$. same antenna. 60ft long wire. same receiver (r.f., f.c., i.f.. det., o-p.), and guess what I heard-nothing. absolutely nothing!
J. R. Hunt (London) has an R1155 into a 67 ft long wire, 15 ft high. and reports between $0700-$ 0905: 9GIDY. W3KFQ. W2LOY, WA2KIC. ZL 3OP, VK 6 MM , all on $14 \mathrm{Mc} / \mathrm{s}$. P. D. Coull (I.ittestone. Kcnt) reports ( $14 \mathrm{Mc} / \mathrm{s}$ ) VK3VJ, VS9MG: at 1600 VE8MI.. VE8RG: at 1800 OX3MN at 1530 and TG9RJ. W2OKM at 1915. On $21 \mathrm{Mc} / \mathrm{s}: 9 \mathrm{G} 1 \mathrm{FC}, \mathrm{ZB} 1 \mathrm{RM}$ at 1030 and WIONK at i410. The gear-a Codar CR66 and a 60 ft long wire.
M. Woollin (Leeds) uses a modified 19 set, an RF24 unit and a 100 ft long wire. This little set-up raked in KA2RG. VK8KK. VK2JZ on s.s.b. at 1540 on $14 \mathrm{Mc} / \mathrm{s}$, while on $21 \mathrm{Mc} / \mathrm{s}$ he really twists the knife into your poor scribe with ZC4MO, 905Y'L. 905PN, ZSIAB, SUIDL, VS9APT, 9CilDM, and all on a.m.. phone ton.
D. Howarth in Bolton, Lancs., uses a PCR 3 receiver and a 20 ft vertical. However, he also confesses to using a long wire with a 6000 resistor at the far end. (The wattage is not stated, so don't write in and ask!) Fourteen Mc/s produced SPSACD, OEGUX, OY7S, OEIHKW, CTILX, W8RRT. OHIWI. FSKE. OEIGE, OHSNQ, EABNA.

Down in wild and woolly Wales stations are coming in very well according to M. Carter, whs listens at Colwyn Bay. On 14Mc/s VK2KM (s.s.b.) and $21 \mathrm{Mc} / \mathrm{s}$ KV4CX (a.m.). W6BMG (s.s.b.) received on a 19 set Mk. 3 with a Labgear front end converter. A recently purchased H.R.O. Jumior. also with a pre-selector, brought in VKil.N. WAGZZ (s.s.b.). ZI.IAY (a.m.) and $K A 7(L H$ (s.s.t.). The twist in the end of this story is that the antentat is an indoor dipole and the house is not more than 50 ft a.s.l. Flat-dwellers proceed to the loft immediately with 33 ft of wire? Incidentally. Michael savs that most Pacific DX was received between 07.30 and 08.30 hours before going to school or 07.30 to 1000 hours at weekends.

Bornard Hushes. BRS2590) (Worcester), runs an Edddustone 840C" and a Codar preselector with a choice of either a 20 m dibole or a 70 ft lone wire. With this set-up his 20m lng is: 5ZHAO. -continued on page 969

# BOOKS REVIEWED 

品RADIO SERVICING ("Teach •Yourself" Books), by L. Butterworth.

- Published by The English Universities Press Led.
. 257 pp. with index. $7 \frac{1}{d i n}$. $\times 4 \frac{1}{2} \mathrm{in}$. Hard cover. Price 10s. 6 d . WHE "Teach-yourself" books have come in for a lot of banter. "How to be a physicist in six easy lessons". chaffs the comedian. But, in fact, a remarkable amount of useful information is contained in these 19 chapters, and the student with some access to practical workshop (or kitchen table) facilities would find this book a very inexpensive course on the subject.

The emphasis on the practical approach is evident from the outset. There is no opening gambit of long-drawn theory but an immediate attack on diágrams, soldering and simple measurement, leading to a second chapter which manages to slip a good deal of theory into a practical discussion of resistance and resistors. This is the technique : throughout the book: the author demonstrates his theory by practical exercises, much as the lecturer would at technical college. Each chapter * has a terminating list of exercises.

After resistance, series and parallel circuits, a.c. is dealt with, in a masterly chapter only a dozen pages in length, yet managing to bring in sine and square waveforms, frequency, measurement and even a description of the oscilloscope. Capacitors, inductances and transformers follow, with the usual protracted electromagnetic theory chopped down to the necessary minimum.

Chapter 7 deals with rectifiers, and this leads naturally to power supplies and thence to amplifiers, bringing in various valves and their characteristics and treating transistors as a natural parallel. Not, as so often, an adjunct to the business of radio servicing. Chapter Ten takes the reader straight from the triode amplifier to the junction transistor a.f. amplifier.

The next chapter. The Output Stage, reverts again to valve operation before leading to the rather special field of output transistorised stages. Here, there is a tendency to gloss over the subject with a remark: "A number of other arrangements of output transistors may be encountered, However, the student who has progressed thus far with the aid of this book should have little difficulty in developing his reading along sound lines.

After a chapter on low frequency amplifiers there is a surprising reversion to tuned circuits and oscillators, and thus to transmission and reception, r.f. circuits, selectivity, bandpass circuits and an
introduction to alignment entering at this point, before a discussion of the superhet receiver itself. The book is rounded off with a chapter on workshop equipment and methods, which is, for a work of this nature, quite inadequate. The essence of radio servicing is the alliance of practical testing with theoretical knowledge. which the remainder of the book most admirably presents. This final chapter could have been much more comprehen-sive.-H.W.H.

## 三 GUIDE TO BROADCASTING STATIONS-14th EDITION <br> Compiled by the staff of "Wireless World" magazine. Published by liffe Books Ltd. $128 \mathrm{pp} .7 \frac{1}{6} \mathrm{in} . \times 4 \frac{3}{4} \mathrm{in}$. Price 5 s.

TTHE greater part of the contents of this extremely useful reference book appear under two main sections: Long and Medium-wave European Stations, and Short-wave Stations of the World, and these two headings almost describe the book's coverage of the subject sufficiently to make further comment superfuous. To elaborate. however, each section gives a comprehensive list of stations in order of frequency, also listing country of origin, power in kilowatts and the wavelength equivalent in metres. In the list of world s.w. stations, call signs are also given, while in the European list. channel numbers are quoted. Each section also provides a separate geographical list of the stations.
The book contains one other list which gives the international allocation of call signs by countries. A short list of the s.w. broadcasting bands precedes the short-wave section and in this section also, stations operated by "extra-territorial" organisations (e.g. British Forces Network, Voice of America) are indicated by abbreviations.

There remains one other contents item to be mentioned, at which can be levelled the only unfavourable comment the book as a whole warrants. A map showing the broadcasting regions of the world in the centre of the book, which would normally have been of much value and interest, is rendered indistinct through the use of poor quality paper. However, as already said this was the only adverse criticism of the book and this, when considered against the reasonable price can amount to no very great objection. Otherwise this book must be considered excellent value to the short-wave listener and the DX-er: cheap enough to discard for a new one as each revised edition appears.-P.R.R.
> "PRACTICAL WIRELESS" INDEX
> Indices for "Practical Wireless" Volume 39 (May 1963 to April 1964) are now available from Post Sales, George Newnes Ltd., Tower House, Southampton Street, London, W.C.2., price. Is. 3d. including postage.

## PART 5-FEEDBACK IN TRANSITOR CIRCUITS

## Understanding SEMICONDUCTORS <br> BY LESLIE MOORE

## CONTINUED FAOM PAGE 890 OF THE JANUARY ISSUE

ANY natural oscillation tends to obey a sine law unless some interference or other, is used to distort that oscillation. Non sinusoidal oscillators exist which produce waveforms that may be square. saw tooth, staircase etc.; these are known as relaxation oscillators. One relaxation oscillator called the "astable multivibrator" or the " free running multivibrator", a fairly common cir-


Fig. 33: A C/R series circuit.


Fig. 34: A graph of the voltage "growth" across a capacitor in a C/R circuit.
cuit throughout the electronics field, produces an almost perfect square wave.

An important factor in the design of the free running and other multivibrators, is that of the charge and discharge characteristics of a capacitor in a d.c. circuif.

Fig. 33 shows a scries circuit including a capacitor C. resistor R. a battery of potential $V$ volts and a switch $S$. When the switch is closed an instantaneous value of current would flow equal to $R / V$ Amps. Current flowing in the eircuit "charges" the capacitor, the charge acting in opposition to the flowing current: the value of charging current then decreases, the voltage across $R$ will decrease so a potential will appear across $C$.

Taking this series circuit at face value we could assume three things:
(i) At the first instant there is voltage drop across the capacitor.
(ii) As time passes the capacitor charges hence causing a voltage drop across itself.
(iii) We could assume that finally no current would flow and hence no voltage drop across the resistor.
If the growth of potential drop across the capacitor were drawn on a graph with respect to time the result would be as in Fig. 34. This growth of potential obeys an "exponential" law. The bottom of the curve appears to be linear.

Suppose the rate of change of voltage across the capacitor at the beginning of the operation were sustained until a voltage equal to that of the supply voltage were reached. The time in which this operation would take place is known as the time constant. As the actual rate of change obeys an exponential law, in the first time constant the voltage rises by approximately $63 \cdot 2 \%$ of the supply voltagc. If the supply voltage were 100 V then after one time constant 36.8 V would appear across $R$. After a second time constant $63.2 \%$ of the 36.8 V would further be dropped across C. The process is unending; theoretically current never stops flowing in the circuit, but in practice after a few time constants have passed any current fowing can be neglected-depending, of course, on the requirements of the circuit.

The time constant of a circuit is the product of capacitance and resistance.
E.g. If an $0.1 \mu \mathrm{~F}$ capacitor were in series with 2 $1 \mathrm{k} \Omega$ resistor the time constant would be:

$$
\begin{aligned}
0.1 & \times 10-{ }^{6} \times 1 \times 10^{3} \text { seconds } \\
& =0.1 \times 10-^{3} \text { seconds } \\
& =0.1 \text { milliseconds }(\mathrm{mS}) .
\end{aligned}
$$

A capacitor discharging across a resistive load also loses its voltage exponentially.

The circuit in Fig. 35 is known as a differentiating circuit". If a square wave were to be applied


Fig. 35: The differentiating circuil.
across the input the output wrould be a function of the combination of capacitive and resistive values. Suppose the time constant is very short compared with the time of existence of any one pulse, the capacitor would be able to charge almost fully to the applied voltage. Fig. 36a shows the output of such a circuit.

A square wave applied to the differentiating circuit with a long time constant compared with the wave time length would produce an output shown in Fig. 36b: this is because the capacitor is not allowed sufficient time to charge fully.

The circuit of a free running multivibrator is shown in Fig. 37. R1 and Trl are in series as also are R4 and Tr2. Because these two series circuits are not quite exactly the same in resistive values due to tolerances etc., one transistor may conduct slightly more current than the other. Assume that Trl conducts a higher value of current than Tr2, more voltage will be dropped across RI than R4,

(a)

(b)

Fig. 36: (a) The output waveform of a differentiating circuit with a short time constont. (b) The output waveform of a differentiating circuit with a long time constant.
then Trl collector will be more positive than Tr 2 collector. In the first instant of applying the supply voltage, the voltage across R1 will also be seen across R3. The base of Tr 2 will be almost the same potential as the emitter, therefore current conduction through $\operatorname{Tr} 2$ will reduce, hence the collector of $\operatorname{Tr} 2$ will rise towards the negative supply voltage. The base of TR1 also rises towards this value and $\operatorname{Tr} 2$ will continue to conduct even more current. This action continues until Trl is conducting its maximum value of current (determined by the value of R1) and $\operatorname{Tr} 2$ is cut off. The capacitors Cl and C 2 begin to charge as soon as current flows to their plates, therefore the base of Trl begins to hold a positive going potential and $\operatorname{Tr} 2$ a negative going one. A point is reached where conduction through Tr 1 and Tr 2 is switched due to the changing base potentials. The switching between Trl and Tr2 will continue because of the action of capacitors C1, R3 and C2, R2. The values of C1, C2, R2 and R3 are important as their time constants determine the circuit's switching frequency.

The output of the circuit is taken from any one of the transistor collectors. If a fairly good square wave is required the value of the output capacitor, $C$, and the resistive value of the load should form as long a time constant as possible.

The frec running multivibrator is often used as a signal generator as there are very few components required and can be built very inexpensively.

## Conclusion

Semiconductors have been seen to have a variety of uses; as a rectifier enabling the conversion of alternating currents into direct current, as an amplifier in the form of a transistor, as sinusoidal and square wave generators, they can be used as switches to half-pulse frequencies-these are only a few of their applications.

Numerous forms of semiconductor devices exist, the majority of which have not been mentioned in this serres, they include the tunnel diode making possible the solving of problems at ultra high speeds. the variable capacity diode which is now used in frequency modulating circuitry replacing cumbersome and inefficient components, the four layer diode and unijunction transistor, each of which can produce saw tooth generations with as


Fig. 37 : Free-running multivibrotor circuit.


Fig. 38: A typical characteristic curve of a zener diode.

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Fig. 39: In this circuit arrangement the zener diode is operating under revers: bias conditions.
few as two other components and a voltage supply, the list is unending.

One particular special purpose semiconductor device of interest to nearly all electronic enthusiasts, is the "Zener diode ".

A normal pn junction diode under reverse bias conditions will begin to conduct current when the voltage applied to the diode is in the region of one to two hundred volts. This voltage is known as the breakdown voltage. A scientist naned Zener examined this effect and because of his work the Zencr diode, which has a breakdown voltage of only a few volts, is now in general use.

Zener diodes are used as voltage stabilisers, several types are in existence, each with a different hreakdown voltage. Typical characteristics of a Zener diode are shown in Fig. 38. From the slope of the characteristic curve between points $A$ and B, it is seen that for a large change in current through the diode. voltage across the diode changes only minutely. The Zener diode is worked on this part of its characteristic as a voltage stabiliser.

A Zener diode is shown in Fig. 39 in series with a resistor, a voltage supply is across the series circuit. Any small change in the supply voltage would cause a current variation in the series circuit. Providing the diode is in that part of its working region equivalent to being between points $A$ and $B$ in Fig. 38, for all practical purposes the voltage across the diode would not change: therefore the difference in supply voltage causes an equal voltage difference across the resistor.

If only a small stabilised voltage was required,


Fig. 40: A voltage stobiliser circuit. employing o zener diode, used as the las: stoge of a power amplifer. The circuit symbol of the zener diade used in these diagrams is only one of several in general use.
say 5 V - a suitable Zener diode would be chosen for a circuit such as that shown in Fig. 39, the output would be taken from across the diode. This is an obvious application of the device, there are many, one of which incorporates the use of further semiconductors, again, as a voltage stabiliser.

The action of a power supply was explained earlier in this series: many circuits require a larger voltage supply than the breakdown voltage of a Zener diode to be stabilised. Fig. 40 shows the type of stabiliser that may be used as the last stage of a power supply.

The stabiliser has two purposes.
(i) to remove any ripple resulting from insufficient smoothing.
(ii) to give a stable d.c. output for any variation of mains voltage to the power supply.

Tr 1 is acting as a variable resistor in the form of an "emitter-follower" circuit. D produces a constant emitter bias for Tr2. If the collector voltage of Tri were to change. the emitter would try to follow that change, but in doing so the base hias of Tr 2 alters. The conductance of Tr 2 changes as a result. The change in emitter-collector voltage of $\operatorname{Tr} 2$ is directly fed back to the ba:e of Tr1. The conductance change of Trl due to this compensates for the change in Trl collector voltage and the emitter of Trl stays at a steady potential.

The stability of this circuit is due to $D$ as it holds the emitter of $\operatorname{Tr} 2$ at a steady bias voltage.

There are many advantages of semiconductors over thermionic devices. They include, the small working voltages required for their operation, their comparatively small pbysical sizes, no power supply is required for their operation. One disadvantage of domestic type semiconductors (e.g. those found in portable radios) is that they must not be overheated.

Many of the problems encountered in electronics are associated with power supplies, and because of this power supplies have been dealt with fairly fully compared with the other subjects covered in this series, but the subject is not quite complete.

It is quite often required to have nore space on a chassis to make minor alterations-a method of providing that space, if not already done, is to convert the thernionic rectifier into one using pn junction diodes. A small mounting on the mains transformer would be sufficient to hold the rectifying circuit. as diodes now exist with ratings of "peak inverse voltage" of 800 V and "peak forward current" of 0.5 A and are only as large as a $\frac{1}{W} \mathrm{~W}$ resistor.

Semiconductors, although they appear to he extremely complicated. have much to offer-there are numerous types doing many operations, there are also many publications explaining in vivid detail the uses of semiconductors. I hope this series has helped you understand them.

# TRANSISTOR RECEIVER SERVICING 

STEP-BY-STEP

EIXPERIMENTERS and enthusiasts brought up on valve equipment sometimes have difficulty in switching their thoughts to transistor circuits and the relatively low voltages that these work with. When dealing with transistor equipment, the author finds it best to forget valve techniques altogether and look upon the transistor circuit as something new.

Although there are several parallels between valves and transistors, as brought out in the classic descriptions of transistor operation, there are often times during a fault tracing exercise when a transistor fault condition has no resemblance whatever to a valve fault condition.


This can complicate matters if one has stuck too rigidly to valve/transistor parallels.

With the in mind, this article is concerned not with how transistors and their circuits work particularly, but what to do when a transistor circuit ceases to function. Fortunately, a transistor is a simple device to work with since it has only three electrodes, these being called the base, the emitter and the collector, as shown on the transistor symbol in Fig. 1.

## by Gordon J. King

There are two basic types of transistor, the $p-n-p$ type which is used extensively in all transistor radios and small amplifiers and the $n-p-n$ type which, until comparatively recently, was rarely seen in domestic equipment. However, mixtures of both types are now finding their way into the new radios and amplifiers, particularly in the output stage where their use together forms the so-called "complementary amplifier".
This type of output stage has the great advantage of requiring neither a driver transformer nor an output transfoimer, the speaker being connected direct to the transistors.

## P-N-P and N-P-N Polarities

Symbols $p-n-p$ and $n-p-n$ are shown at (a) and (b) respectively in Fig. 1, and one important point to note at this juncture is that the collector of the p-n-p type is connected to supply negative while the electrode of the $n-p-n$ type is connected to supply positive. If something is done during a servicing operation to cause a reversal of this potential, then it is likely that the transistor will burn out or be impaired considerably in efficiency.

When working as an amplifier, a transistor passes current through its collector/emitter circuit, but the current in the emitter circuit is very slightly higher than the current in the collector circuit since the enntter circuit carries also a little base current.

Collector current will not flow unless a little base current is flowing, and in order to satisfy this latter requirement the base of a p-n-p transistor must be connected to a negative potential while that of an $n-p-n$ transistor must be connected to a positive potential.

Both the collector and base potentials are relative to the emitter. as shown in Fig. 2 at (a) for p-n-p transistors and (b) for n-p-n transistors..

The input signal may be applied either at the emitter or the base and extracted from either from the collector or emitter. The three resulting methods are shown in Fig. 3. That at (a) is the "common-emitter", at (b) the "common-hase" and at (c) the "common-collector". Note that the common electrode in each instance is that which carries both the input and output signals.

## Circuit Configurations

The common-emitter circuit has the highest power gain and is the most frequently used. The input impedance is medium and the output impedance medium to high.

(a)

(b)

Fig. 1: Transistor circuit symbols; (a) $p-n-p$ and (b) $n-p-n$. Note change in direction of emitter arrow head.


Fig. 2: Transistor polarity; (a) $p-n-p$ and (b) $n-p-n$ Voltoges are relative to emitter.


Fig. 3: Transistor configurations; (a) common-emitter, (b) commonbase and (c) common-collector.

Transistor amplifiers can be subjected to both negative and positive feedback by applying from the output out-of-phase or in-phase signals to the input. In the latter respect, therefore. coupling for positive feedback turns the amplifier into an oscillator. the frequency of which is controllable either by CR or LC circuits in the usual manner.

The most usual method of applying the base and collector potentials to an amplifier is shown at (a) in Fig. 4. The potential-divider R1/R2 allows the base to be biased from the same supply source as feeding the collector. It also stabilises the basc bias.

The emitter resistor R3 also serves to stabilise the transistor against an increase in collector current which can otherwise arise due to temperature increase, this being a characteristic of the transistor.

With R3, if the collector current rises, the emitter current also rises and the increase in volts drop across the resistor pulls back the base voltage and thus reduces the collector current.

## D.C. Feedback

This gives d.c. feedhack, but a.c. feedback is avoided in the common-emitter configutation by bypassing the emitter resistor with a capacitor having a low reactance at the signal frequency. In the common-collector configuration the basic stage is subjected to full negative feedback, since the emitter resistor cannot be bypassed as it represents the signal load.


Fig. 4: Base biasing and stabilising arrongements; (o) bose potential-divider $R 1 / R 2$ (this is the most popular and effective), (b) with o single resistor to the supply line (stability poor) and (c) single base resistor returned to collector (improved stability).

Let us suppose that we have a single stage amplifier, as shown in Fig. 5, that fails to work. What should be done first? The obvious thing, of course, is to ensure that the supply voltage is present. A voltmeter with a fullscale deffection of about 25 V connected as for Test 1 in Fig. 5 will soon prove this possibility.

## Open-circuit Collector Lood

If the supply is correct, the negative lead of the voltmeter should be transferred to the collector as shown by Test 2 in the diagram. The load whether it be indactive or resistive, should have continuity, so this time we should read sométhing a little

Fig. 4(b) shows an alternative method of biasing the base by the use of a single. higher value resistor between base and the supply line. This. unfortunately, suffers from the misfortune of rather poor d.c. stability. A better arrangement is shown in Fig. 4(c), where the base resistor is returned to the collector end of the load instead of to the supply line.

Now armed with the basic facts of transistor circuits, we can investigate a few fault conditions.


Fig. 5: Four mojor voltage tests which avoid breaking the circuits to insert a current meter. These are fully described in the text.
less than the supply voltage, depending upon the resistance of the load and on the current in it. If there is no collector voltage, then the fault is due simply to an open-cirquit load.

If the collector voltage is about the same as the supply voltage and the load is resistive, this would indicate very little or no collector current. On the other hand, if the volts drop across the load is excessive one may conclude that the collector current is abnormally high. The best way of checking the collector current is to measure the voltage across the emitter fesistor, as shown by Test 3 in the diagram.

If, for instance, the emitter resistor is $1,000 \Omega$ and 2 V is measured across it, then we know from Ohm's law (current equals the voltage divided by the resistance) that the emitter current is 2 mA .

The collector current is a little less than this calculated value, of course, since base current also flows in the emitter resistor. However, as the base current is only a matter of $100 \mu \mathrm{~A}$ or so, the current difference between emitter and collector can usually be ignored when making simple tests.

If there is zero volts drop across the emitter resistor, then either the transistor is faulty or there is no base current (always remember that there must be base current to produce collector/emitter - current of measurable magnitude). It could - happen, of course, that the emitter resistor bypass ?f.capacitor has a short-circuit, but this is unlikely.

How, then, would one ascertain whether the trouble lies in the transistor or in the base bias circuit? One way would be to test the transistor either on a tester or by substitution. This is a timeconsuming exercise and is not always easy on a densely-mounted printed circuit board.

There is also the danger of ruining the transistor or nearby components by excessive and prolonged heat from the soldering iron. By far the best idea is to ensure that the transistor is faulty before contemplating its removal from the circuit.

## Final Test

If we have followed all the tests up to date, a final test, shown as Test 4 in Fig. 5, should verify the imatter one way or the other. Here we are measuring base voltage pure and simple relative to the emitter. The voltage will be very small, for it needs only a fraction of a volt to give a base current of $100 \mu \mathrm{~A}$ or thereabouts.

However, if there is a deflection on the test meter, one can be reasonably sure that the lack of collector current is due to transistor fault, as distinct from a fault in one of the associated components.

There are two impartant points. One is that the $\$$ polarities given in Fig. 5 are for a p-n-p transistor. With a n-p-n transistor, the tests would be the same but the meter polarity would need to be reversed. The second, concerns test meter application and sensitivity.

Voltage and current transients can spoil a good transistor, so to avoid such happenings the meter should be connected and. removed for the various tests with the circuit disconnected from the power supply. Also, the very small voltages, especially emitter and base voltages, call for a test meter with a full-scale deflection not exceeding about $2 \cdot 5 \mathrm{~V}$. A f.s.d. of IV is to be preferred.

It is important that the meter has a sensitivity of at least, $10.00052 / \mathrm{V}$, otherwise the loading effect of the connected meter will disturb the circuit voltages and probably lead to incorrect interpretations.

So far we have considered very low or zero voltage a ross the emitter resistor. What do we do if the voltage here is abnormally high? It is true that this condition does arise, and an abnormal voltage on a small transistor would be considered in excess of that produced by a maximum current of 2 mA .

For example, if 4 V is measured across a $1,000 \Omega$ emitter resistor, the current would be 4 mA . which is abnormally high for a small signal transistor.

## High Base Current

This condition can be brought about by two major factors. One is excessive leakage in the transistor itself and the other a too high value of base current. Thus, it would be best to check first on the base voltage, as in Test 4, Fig. 5.

Make sure that the meter clip is actually contacting the base wire, for should the base end of R1 develop a dry joint on the printed wiring, the meter connected to that point would register zero volts. when in actual fact the base of the transistor proper would be highly negative. This is because the potential-divider effect would be destroyed (with R1 disconnected from the base wire) and base current would be limited only by the value of R2. It would thus rise to a far higher value than is normal when held down by the connected R 1 .

This, incidentally, is a frequent cause of high emitter current, not so much by RI actually going open-circuit but by a dry-point developing at one end.

## Copacitor Leak

It sometimes happens that the base and collector potentials are affected by a leak in the input or output coupling capacitors-C1 and C2 respectively in Fig. 5. This can be proved easily enough simply by disconnecting one end of the component and measuring the electrode voltages again.

A significant difference between the original reading and that made with the coupling capacitor disconnected, should lead to replacement of that capacitor. In audio circuits electrolytics are used extensively in these positions, and these tend to leak with age or when connected round the wrong way.

With p-n-p transistors, the negative side of an interstage coupling capacitor should be connected to the collector of the preceding transistor and the positive side to the base of the following transistor. The polarity is reversed, of course, with $n-p-n$ transistors.

## Self-oscillating Mixer

Basically, the majority of transistor stages can be resolved to that in which the tests are shown in Fig. 5. Fig. 6 gives the basic circuit of a self-oscillating mixer, as is found in most transistor portables. This has all the basic features. We have the base potential-divider comprising R1 R2, the emitter

resistor $R 3$ with its bypass and the collector load which, in this case, consists of L1 and the primary of the i.f. transformer T1.

This, then, amplifier-wise, is a common-emitter circuit with the aerial signal applied to the base. Both the collector and emitter are loaded by Li and L2 of the oscillator coil respectively. Positive feedback is thus applied from the collector to the emitter, causing the stage to oscillate. The difference-frequency (that is, the i.f.) is developed across the i.f. transformer which also acts as a second load in the collector circuit.

The transistor is biased by R1/R2 to approximate class $A$ conditions, but when oscillation commences the transistor is driven towards class $B$ conditions. This is because the base/emitter junction of the transistor rectifies the oscillatory signal.

## Oscillator Test

Test 3 of Fig. 5 would give a slightly higher reading when the stage is oscillating than when it is quiescent. Here. then. is a test for oscillation. If there is a small decrease in voltage across the emitter resistor when the oscillator gang section of the tuning capacitor is shorted, then one can be sure that the stage is oscillating, at least.

By following the reasoning given in this article

Fig. 6 (left): Basic self-oscillating mixer, as used in the majority of transistor portables.


Fig. 7 I The basic class $B$ output stage.
almost any transistor circuit fault can be diagnosed without having a great deal of experience in the servicing of transistor equipment.

Output stages have the same basic set-up, but here it is usual to find them biased towards class $B$ conditions. Fig. 7 reveals a typical class B output stage, in which R1 and R2 form the potentialdivider which biases both transistors and R3 is the common-emitter resistor. The bases are biased not exactly to cut-off, since this produces an unpleasant noise, called "crossover distortion ", but to a small quiescent current.

Note that the emitter resistor is not bypassed in. this circuit, and a thermistor may be employed instead of an ordinary resistor to enhance the' thermal stability. The voltage across this component would normally be very low at zero signal due both to the low resistive value and the small quiescent current. However, the current rises to high values, as does the voltage across the emitter element, with increase in sound output.

## Distortion

Excessive distortion can be caused by (i) unbalance in the transistors, (ii) incorrect hiasing (due, for instance, to alteration in value of RI or R2) and (iii) low battery voltage.

# P.W. Film Show <br> as a reader of "Practical Wireless"... 

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IN the December, 1964, issue of Practical Wireless the author described the construction of a short wave superhet recciver particulaty suited to the beginner. The receiver has now been in use for several months and has given an extremely good account of itself. In the interests of simplicity several desirable features were omitted from the original receiver. The chief
among these were a stage of radio frequency amplification prior to the frequency changer and a beat frequency oscillator (b.f.o.) to enable continuous wave (c.w.) transmissions to be received.

The author has now incorporated these features in the original design and achicved a worth-while increase in performance. For the benefit of the


Fig. I: The preselector unit circuit diagram.

[^3]beginner and particularly for those who built the receiver a full description of these adjuncts is given.

## The Radio Frequency Amplifier

For peak performance over $10 \mathrm{Mc} / \mathrm{s}$ or so, a stage of amplification at the signal frequency is a desirable feature in any receiver. Not only does it improve the signal-to-noise ratio but also a distinct reduction in second channel interference is obtained. For a brief discussion of this effect the reader is referred to the original article in the Decenïber issue. Noise in a receiver arises not only from man-made and natural interference but also from the receiver itself. Generally most of the noise is contributed by the first stage of the receiver and then principally by multi-electrode valves such as the frequency changer. Since this noise in the preliminary stage is amplified by the later stages then in effect only this early noise is of importance. A weak signal presented to the mixer would obviously be largely masked by the noise arising in this stage. An improvement in the signal-to-noise ratio would therefore be
objective an increase in selectivity. Although the author has freely used both terms in the text the distinction should be borne in mind.

The r.f. amplifier described in this article enhanced the performance of the original receiver, particularly on the 14 and $30 \mathrm{Mc} / \mathrm{s}$ bands. With the amplifier in circuit hitherto weak signals were easily resolved.

## Circuit Description

The circuit diagram of the instrument is given in Fig. 1. As will be seen, the circuit is quite simple and consists of a single-ended semi-variable mu pentode which is a 6SG7, used in conjunction with the Denco series of octal-based plug in coils. Tuning, of the unit is by means of a 300 pF variable' capacitor Cl . Since this is the value of the variable capacitor in the main receiver, excellent agreement between the unit and receiver dial readings is obtained. The gain of the amplifier is regulated by VRI, which is a wire-wound component. The r.f. output is fed via C4 to the output socket SK2.

## ER'S S. W. SUPERHET

## H. WEBSTER

obtained by a stage of r.f. amplification prior to the mixer. Although such a stage is correctly called a preamplifier. frequently it is referred to as a preselector, the two terms being regarded as synonymous. A preamplifier increases the sensitivity of a receiver, while a preselector has for its

A variable mu 6SK7 may be used in place of the 6SG7 with a slight loss in gain. The-author also found that with VR1 set at maximum both the 6AC7 and 6SH7 typés gave an extremely good performance. Both these types are, of course, nonvariable mu.


Fig. 2: Above-chossis drilling detoils for the preselector.


Leth hand side chassis member looking at rear of chascse


Fig. 3: Drilling detalls of the chassis members.


Fig. 4: Dimensions of the preselector front panel, showing layout adopted by the author. This may also be seen from one of the heading illustrations on page 964, the other photogroph being of the b.f.o. unit mounted on the receiver chassis.

## Constructional Details

The unit is constructed on a $\operatorname{Sin}, x \sin . x 3 i n$, chassis and the front panel is a nominal 9in. $x$ 9in. chassis top plate taken from the same chassis range. Full drilling details for the chassis and panel are given in Figs. 2, 3 and 4. It will be noted that no dimensions are given for the fixing of the variable capacitor. The author recommends that the holes for this component are drilled after the dial unit is fixed to the front panel. A further important point also concerns the variable capacitor mounting. The panel drilling details are applicable to the specified capacitor, which is a Jackson type E1. If another make of capacitor is used slight alteration of the dial height may be necessary. Careful attention to this point must be made before drilling the front panel.

Although tuning of the unit is relatively easy an Eddystone slow-motion dial was chosen for two reasons: first to provide calibration points identical with those on the main receiver and secondly to match up with the appearance of the main receiver. The drilling details of


Fig. 5: Underchassis wiring diagram of the preselector.
the front panel are applicable to this dial.
When drilling of the front panel and chassis is complete the unit is temporarily assembled. The flexible coupler is placed on the dial shaft and the variable capacitor offered up to the coupler. Ensure that the capacitor shaft is in exact alignment with the dial spindle. Spacing washers under the mounting feet may be found to be necessary to achieve an accurate fit. When alignment is satisfaclory the chassis holes are marked by pricking through the capacitor mounting feet. The panel and chassis may now be dismantled and the holes for the variable capacitor bored at the appropriate points. The fitting and wiring of components is facilitated by working on the dismembered chassis. The ceramic coil and valve holders are mounted, making sure the holders are correctly orientated. The mounting of the screening can for Li, L2 is described in the leaflet which accompanies the coils.

Great care is necessary when cutting the screcning can base, since slight distortion of the thin gauge metal will cause an unsatisfactory fit. It is recommended that cutting of the base is done with a sharp razor blade of the singled-edged type.

As an aid to easy coil changing the author found that a preliminary polish of the threaded portions of the can and hase with metal polish greatly facilitated the removal of the can. The remaining components are assembled as shown in Figs. 5 and 6. Since the frequency stability of the unit is important, all wiring should be carried out in heavy gauge copper wire. Wiring of the gain control, toaxial sockets and high frequency choke is completed when the chassis is assembled. A short piece of copper braid stripped from a piece of coaxial cable is used to earth the frame of the variable capacitor to the fixing screw on the holder of L1, L2. Do not rely on the fixing feet of the capacitor as being a sufficient earth.

When wiring is complete a range 3 coil is plugged into the holder and a check made of the resistance between h.t. nositive and chassis This, of couse, is a wise precaution before connecting any new equipment to a power supply. A slight reading will be obtained initially due to C2 and C5. After a little time the meter should register practically infinite resistance. The unit is connected to the receiver power pack. In the authors unit an additional octal valve holder was

Fig. 6 (right): Layout and wiring abive-chassis.
fitted to the power unit and the amplifier power leads terminated in an octal plug. The same wiring convention used in the main receiver was adopted for the wiring to the plug and socket. A short length of coaxial cable which terminates in a coaxial plug at each end is connected between SK2 and the main receiver aerial socket. A set of range 3 coils is plugged into the main receiver and the power unit switched on. When steady conditions are obtained the following voltage readings should be obtained. The author used a $10,000 \Omega / \mathrm{V}$ meter set to the $1,000 \mathrm{~V}$ range for h.t. measurement and the 10 V range fot bias voltage measurement.

Voltege Readings at 65G7. Gain Control Maximum:

| Anode <br> Pin 8 | Screen <br> Pin 6 | Cathode <br> Pins 3,5 |
| :---: | :---: | :---: |
| 205 V | 110 V | 2.1 V |

The aerial is connected to SKI on the amplifier and the amplifier and receiver dials set to $85^{\circ}$. After setting the gain control to maximum the iron core of L1, L2 is adjusted until a marked increase in noise level occurs. The core of L1, L2 may now be secured either with a blob of sealing compound of by running a 6 BA nut down the free end of the thread. If this latter device is adopted the core must be held securely by a screwdriver while the lock nut is tightened.

The remaining coils (ranges 4 and 5) are dealt with in a similar fashion. It will be found that
 is very simple. Over a small range of frequencies such as an amateur band the amplifier may be peaked in the centre of the band and fine tuning done on the main receiver. Slight adjustment of the amplifier will only be necessary at the band edges.

With VR1 set for maximum gain it may be found that oscillation occurs on the highest frequencies. A slight reduction in the settin of VR1 will therefore be necessary. The author found that the unit was quite stable and no trouble should be experienced on this score. If the constructor encounters this effect a $10 \Omega$ resistor wired in series with the grid leak of the 6SG7 should provide a cure. This resistor should be mounted as near as possible to the grid pin.

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# "WALKIE-TALKIES" A CLARIFICATION OF THE POSITION 

FTOLLOWING our recent editorial comments on the subject of amateur radio pirates, which touched upon the question of the various walkie-talkie transceivers being sold in this country, we feel it would be helpful both to readers of this magazine and to advertisers to print below a communique issued by the GPO. This makes clear any doubts which may exist concerning the use of such equipment.

## "WALKIE-TALKIES"

Any use of radio communication in the United Kingdon must be licensed by the PostmasterGeneral, and the Post Office has recently conducted successful prosecutions for the unlicensed use of "walkie-talkie" radio equipment operating on frequencies around $27.12 \mathrm{Mc} / \mathrm{s}$. It may not be generally known that the Postmaster-General will not license the transmission of speech on these frequencies as the band has been designated for industrial, scientific and medical use.

Transistorised transmitter/receivers are also available working on slightly higher frequencies, c.e. $28.5 \mathrm{Mc} / \mathrm{s}$, but this frequency falls in one of the bands assigned for use by licensed radio amateurs. Refore an amateur licence is granted the applicant has to pass the Radio Amatcur (Technical) Examination and the Post Office 12-words-per-minute Morse sending and receiving test.

Dealers would therefore be well advised to bear in mind, when offered "walkie-talkie" equipment using a frequency of $27.12 \mathrm{Mc} / \mathrm{s}$ or thereabouts, that the Post Office will not issue licences to purchasers of this equipment and any use, including that for demonstration purposes, without a licence would be contrary to the law. Furthermore, equipment operated by licensed amateurs in the $28.0-29.7 \mathrm{Mc} / \mathrm{s}$ band must be capable of operating within the terms of the Amateur (Sound) Licence-for example, a satisfactory method of frequency stabilisation must be employed in the sending apparatus and frequency measuring equipment must be provided capable of verifying that the sending apparatus operates within the frequency band of $28.0-29.7 \mathrm{Mc} / \mathrm{s}$.

Facilities are available for private mobile radio services using speech in the $80,160,170$ and $460 \mathrm{Mc} / \mathrm{s}$ bands and suitable equipment which conforms to the relevant Post Office specification can be obtained through the trade,

It is also worth mentioning that most of the ex-Service mobile radio equipnent on the market operates on frequencies not available for private mobile radio services in the United Kingdom and could not be adapted to operate satisfactorily within the narrow v.h.f. channel available. The use of such equipment is therefore not permitted by the Post Office on private mobile radio services.

## On the Short Waves

## -continued from page 953

9X5GG / P, K4ASR / MM. WA2EZJ /M, W1EQP/4, VK2AHT, OX3JD, ZB1RS, EL1H, OX3LP, K1JTN/P4, 5H3JR, ZB1RM, ZS6U, CR7CO, all s.s.b. or a.m.
A. H. Trickey (Bristol) has an R208 and on erecting a 20 m dipole found he only had a 15 m garden, so a "bit" was lopped off each end. Although it now resonates at about $17 \mathrm{Mc} / \mathrm{s}$ and is bridged across for other bands, $21 \mathrm{Mc} / \mathrm{s}$ proved lively with F5BV, G3OYF, W5PFY/MM, W3UGW, WA4FWN, W3KFK, PY1SZ, W4SUF, G5FS, KV4CX, W1BNH/MM, OH5SM, YO2DZ, UA3KND. W2AWD. K3CMH, W1NRK, K4BIY, W2WLI, W2JY, W1ZER, W2KLN and W4ZLB.
E. Conway (Exeter), using the antenna I described in the November issue, heard PJ2NA and VP7NT in Nassau. M Opie (Dorset) uses a domestic receiver. HMV model 511446 , with an end-fed long wire. A one and a-half hour session one Sunday raised WA4TLI, WA4SAN, W4TUQ. W2BLQ, K1IJF, K3MBF, WA0HHX, all on
$21 \mathrm{Mc} / \mathrm{s}$. While on $14 \mathrm{Mc} / \mathrm{s}: \mathrm{OH} 5 \mathrm{NQ}$, IIPAD, VE3ATU, TF3WJC, W8GZL, W4BBW, WA5KBO. WB2LRN. A week later another period of one and tbreequarter hours on $21 \mathrm{Mc} / \mathrm{s}$ : KV4CX (Virgin 1slands), EL5LA, ON4PH, W8RLT. K2JDW, K3HQM, and on $14 \mathrm{Mc} / \mathrm{s}$ : IIVK, IIRT, OH5RZ, VOIVR, VE2AKZ.

## In General

W1BB is definitely " at it" on 160 m and if you are set on hearing something special on top band then the early mornings are well worth a try,

Gough Island is reported active under a ZD9 call sign and there is a lonely soul in Antarctica signing OR4VN. If you should hear UA1KED don't pass him over as just another Russian station -he's located in Franz-Joseph Land

Contest enthusiasts have the affiliated societies contest on January 23rd-24th and on January 30th31 st. the ether will really be humming for this period with three contests. There's a C.W. contest on 2 m , while at the other end of the scale is the CQWW $1.8 \mathrm{Mc} / \mathrm{s}$ contest. The French REF contest C.W. section is also on for Morse fans, so all in all this period should give SWLs quite a busy period.

## PREPARING Roan <br> - <br>  <br> BRIAN ROBINSON.

## 4: IMPEDANCE, RESONANCE-COUPLED CIRCUITS

### 4.1 Impedonce

Tf a circuit contains not only resistance but also reactance (either capacitive or inductive) the combined effect of the two is called impedance. Although impedance strictly refers to a combination of resistance and reactance, the terms "resistive impedance", "capacitive impedance" and "inductive impedance" are often met. Impedance is represented by the letter $\mathbf{Z}$.

### 4.2 Reactance and Resistance in Series



(b)

Fig. 30-Resistance in series with (o) capacitance and (b) inductance.

When reactance and resistance are connected in series, as shown in Fig. 30, the impedance will be given by-

$$
Z=\sqrt{R^{2}}+X^{2}
$$

-where X is either the reactance of the capacitance in Fig. 30a, or the reactance of the inductance in Fig. 30b.

To find the impedance of the series circuit in Fig. 30a, firstly find the reactance of the capacitance( $f$ in cycles per sec. and $C$ in farads)

$$
\begin{aligned}
\mathrm{X}_{\mathrm{c}} & =\frac{1}{2 \pi \mathrm{fc}} \\
& =\frac{1}{6.28 \times 100 \times 0 \cdot 000002} \\
& =796 \Omega(\text { approx. })
\end{aligned}
$$

Therefore the impedance $Z$ equals-

$$
\begin{aligned}
& \sqrt{R^{2}}+X_{c} \\
= & \sqrt{ } 100^{2}+796^{2} \\
= & \sqrt{ } 643,590 \\
= & 802 \Omega
\end{aligned}
$$

or in Fig. 30b firstly find the value of the inductive reactance-
(f in cycles per sec. and $L$ in henrys)

$$
\begin{aligned}
\mathrm{XI} & =2 \pi \mathrm{fL} \\
& =6 \cdot 28 \times 100 \times 1 \\
& =628 \Omega
\end{aligned}
$$

Therefore

$$
\mathrm{Z}=\sqrt{R^{2}+X L^{2}}
$$

$$
\begin{aligned}
& =\sqrt{ } 50^{2}+628^{2} \\
& =\sqrt{ } 396,870 \\
& =630 \Omega
\end{aligned}
$$

(Remember that in a series reactance/resistance circuit where two or more reactances are connected in series: the total reactance is)-

$$
X=X_{1}+X_{2}+X_{3}
$$

4.3 Reșistance, Capacitance and Inductance in Series


Fig. 31-Inductance, capacitance and resistonce in series.
When resistance, inductance and capacitance are connected in series as shown in Fig. 31, the impedance of the combination will be given by-

$$
Z=\sqrt{ } R^{2}+\left(X L-X_{c}\right)^{2}
$$

For example, in Fig. 31, the value of $X_{C}$ is $1,592 \Omega$ and that of $\mathrm{XL}_{\mathrm{L}}$ is $3,140 \Omega$,

Therefore $\quad Z=\sqrt{ } 100^{2}+(3,140-1,592)^{2}$

$$
\begin{aligned}
& =\sqrt{10,000}+2,396,600 \\
& =1,551 \Omega
\end{aligned}
$$

4.4 Resistance and Inductance or Capacitance in Parallel

(a)

(b)

Fig. 32-Resistance in porallel with (o) copacitance and (b) inductonce.

In the circuits shown in Fig. 32, the impedance is given by the expression-

$$
Z=\frac{R X}{\sqrt{R^{2}+X^{2}}}
$$

-where $X$ is equal to $X_{c}$ in Fig. 32a and $X_{L}$ in Fig. 32b.

In Fig. $32 \mathrm{a}, \mathrm{R}=100 \Omega$ and $\mathrm{X}_{\mathrm{c}}=3185 \Omega$.
Therefore

$$
\begin{aligned}
Z & =\frac{100 \times 3185}{\sqrt{100^{2}+3185^{2}}} \\
& =\frac{318500}{3186} \\
& =99.9 \Omega \text { (approx.) }
\end{aligned}
$$

### 4.5 Impedance applied in Ohm's Law

Impedance, current and voltage can be connected by the form of Ohm's Law

$$
1=\frac{E}{Z} \text { or } E=I Z \text { or } Z=\frac{E}{1}
$$

For example. It in Fig. 32a the applied voltage wad 250 , the current flowing in the circuit would be given

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$$

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$$
\begin{aligned}
I=\frac{E}{Z} & =\frac{250}{99.9} \\
& =0.25 \mathrm{amps} \text { (approx.) }
\end{aligned}
$$

### 4.6 Power Factor

In a circuit containing resistance and reactance, only the resistance actually consumes power. The power factor of a circuit in which an alternating current is flowing is the ratio of the actual power consumed to the apparent power consumed.


Fig. 33-Resistance and capacitance in series with a 250 V a.c. supply.

In Fig. 33 the applied voltage is 250 and the impedance of the circuit is 3186 ohms. The current flowing in the circuit is given by

$$
\begin{aligned}
\mathrm{I}=\frac{\mathrm{E} \quad 250}{\mathrm{R} \quad 3186} \\
=0.0638 \mathrm{amps} .
\end{aligned}
$$

Therefore the apparent power consumed is

$$
\begin{aligned}
W & =E I \\
& =250 \times 0.0638 \\
& =15.9 \text { watts. }
\end{aligned}
$$

Only the resistance actually consumes power however and this is given by

$$
\begin{aligned}
\mathbf{W} & =J^{2} R \\
& =0.0638^{2} \times 100 \\
& =0.41 \text { watts. }
\end{aligned}
$$

factor will be-
$\frac{\text { Actual Power }}{\text { Apparent Power }}=\frac{0.41}{15.9}$

The Power Factor may also be expressed as a percentage-and remembering that for a purely resistive circuit the power factor is 1 (or $100 \%$ )-the power factor in the above example would be $2.6 \%$.

### 4.7 Resonance

The frequency at which theinductive reactance and capacitive reactance are equal, in a radio frequency circuit, is called the resonant jrequency.
$\mathrm{XL}=\mathrm{X}_{\mathrm{c}} \quad$ substitute in the expression, $2 \pi \mathrm{fL}$ for $\mathrm{X}_{\mathrm{L}}$ and $\frac{1}{2 \pi \mathrm{fc}}$ for $\mathrm{X}_{\mathrm{c}}$ and we obtain-

$$
f=\frac{1}{2 \pi \sqrt{L C}} \text { where } f \text { is in cycles per }
$$



Fig. 35-Four examples of coupled circuits.
Four methods of coupling circuits are shown in Fig. 35. In (a) the capacitance C is used as a common feature to both circuits and therefore permits a transfer of power. In (b) the primary circuit only is

## Hints for improved f.m. <br> IETE ME, IROEYAI

NOW that it is becoming increasingly more difficult to use the medium-wave band after dusk, due to the rising whistles and "monkey, chatter" from closely adjacent and "on-frequency" interfering stations. especially at the top end of the band. more and more seribus listeners are turning their attention to the possibilities of achieving improved results on the f.m. (frequency modulation) band.

In the v.h.f. Band II spectrum-from 87.5 to $100 \mathrm{Mc} / \mathrm{s}$ - he BBC transmit at relatively high quality the Home. Light and Third programmes and also regional progammes in some areas. Under normal reception conditions these signals are totally immune from the interference troubles of the medium waves.
the f.m. service have been obtained by a little attention to detail and by one or two simple adjustments, which this article sets out to describe. Firstly. however, it should be remembered that the vast majority of the British Isles is now adequately embraced by the f.m. service, either from main transmitters or from booster or relay stations. Table 1 gives a complete list of such stations. existing and projected.

## Signal Propagation

Secondly, to understand the v.h.f. system. an elementary idea of how the signals are propasated is desirable. Under normal conditions the v.h.f. signals from the transmitting aerial follow a very


Fig. I: Signal propogation. (o) Here the v.h.f.'f.m. signal follows a slightly curved path from transmitter to set, while sky signals pass through the reflective layers. At (b) is shown that at medium frequencies the ground wave hugs the earth over a greater distance while the sky signals are reflected back to earth from the upper atmospheres. Thus, while the range of a v.h.f. signal is limited, medium frequency signals are propagated over far greater distances.

However, a listener going over to f.m. for the first time may be cadly disappointed in the results. for instead of the clear reception expected he may find the programmes badly disturbed by things like crackles from passing cars, by a peculiar warbling note effect marring the tuned programme, and in some cases by a distortion of a kind that leaves the impression of an incorrectly centred loudspeaker speech coil.

These apparent shortcomings, coupled with the possible difficulty of tuning in the required station. may cause the listener to abandon the f.m. service and resort back to the medium frequencies with the " monkcy chatter" and whistles.

In the majority of cases of this nature investigated by the author, greatly improved results on
slight curve over the local surface of the earth. as shown in Fig. 1(a). This means that at distances exceeding the curved path length the receiver is virtually outside the range of the transmitter.

This gives the service area of a v.h.f. station which. rarely, for consistent reception. exceeds 50 to 60 miles from transmitter to receiver, though height of the receiving aerial above sea-level will. of course, have some bearing on this, as would be expected.

The v.h.f. service is thus a "local" service, for consistent use of iisteners only within the range of the service area. This is the reason for the large number of transmitters required to cover the country as revealed in Table I.

Now, the signal propagation at medium

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TABLE

| Station | Main | Relay Booster | Frequency Mc/s |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Ashkirk (S.E. Scorland) |  | * | Light 89.1 | Third 91.3 | Home 93.5 |
| Blaen-plwy (W. Wales) | * |  | 88.7 | 90.9 | 93.1 |
| Divis (Belfast) | * |  | 90.1 | $92 \cdot 3$ | 94.5 |
| Douglas (Isle of Man) | * |  | 88.4 | 90.6 | $92 \cdot 8$ |
| Dover | * |  | $90 \cdot 0$ | 92.4 | 94.4 |
| Enniskillen |  | * | $88 \cdot 9$ | $91 \cdot 1$ | $93 \cdot 3$ |
| Fort William |  | * | 89.3 | 91.5 | 93.7 |
| Haverfordwest (Pembrokeshire) |  | - | 89.3 | 91.5 | 93.7 |
| Holme Moss (Huddersfield) | * |  | 89.3 | 91.5 | 93.7 |
| Kinlochleven |  | * | 89.7 | 91.9 | 94-1 |
| Kirk o' Shotts | * |  | 89.9 | 92.1 | $94 \cdot 3$ |
| Les Platons (Channel Islands) | * |  | 91.1 | 94.75 | $97 \cdot 1$ |
| Llanddona (Anglesey) | * |  | 89.6 | 91.8 | $94 \cdot 0$ |
| Llandrindod Wells |  | * | 89.1 | 91.3 | $93 \cdot 5$ |
| Llangollen (N.E. Wales) | * |  | 88.85 | 91.05 | $93 \cdot 25$ |
| Londonderry |  | * | 88.3 | 90.55 | 92.7 |
| Meldrum (Aberdeen) | * |  | 88.7 | 90.9 | $93 \cdot 1$ |
| North Hessary Tor (S. Devon) | * |  | 88.1 | 90.3 | $92 \cdot 5$ |
| Oban |  | * | 88.9 | 91.1 | 93.3 |
| Orkney | * |  | 89.3 | 91.5 | 93.6 |
| Oxford |  | * | 89.5 | 91.7 | Mid 93.9 West 95-85 |
| Peterborough | * |  |  |  | 94.5 |
| Pontop Pike (Newcastle) | * |  | 88.5 | 90.7 | 92.9 |
| Redruth (Cornwall) |  | * | 89.7 | 91.9 | 94.1 |
| Rosemarkie (N. Scotland) | * |  | 89.6 | 91.8 | 94.0 |
| Rowridge (Isle of Wight) | * |  | 88.5 | 90.7 |  |
| Sandale (Carlisle) | * |  | 88.1 | $90 \cdot 3$ | $\begin{array}{\|ll} \text { Nth } & 94.7 \\ \text { Scot } & 92.5 \end{array}$ |
| Sheffield |  | * | 89.9 | 92.1 | Scot 94.3 |
| Skye |  | * | - | - | 93.9 |
| Sutton Coldfield | * |  | 88.3 |  | $92 \cdot 7$ |
| Tacolneston (Norwich) | . ${ }^{*}$ |  | 89.7 | 91.9 | 94.1 94.5 |
| Thrumster (Wick) Wenvae | * |  | 90.1 89.95 | 92.3 96.8 | 94.5 Welsh 94.3 |
| Wenvae | * |  | 89.95 | 96.8 | West 92.125 |
| Wrotham | * |  | 89.1 | 91.3 | 93.5 |

NOTE-"Relay Booster" indicates a transmitter which derives its signals "off-air" from a main transmitter, and re-radiates them on different frequencies into the "screened" area locally. Stations indicated "Main" may be high or low power transmitters.
frequencies differs substantially from that at v.h.f. as shown in Fig. 1(b). Here one transmitter covers large slices of country because the signal not only follows a curvature closely analogous to that of the earth itself for a relatively greater distance than the v.h.f. signal, but also because the signal rises skywards and is reflected earthwards by " reflective" layers in the ionosphere and troposphere so that great laps of the earth are embraced.

The reflective effect is influenced by time of the day (and might), and after dusk distant stations which do not cause interference during the hours of daylight start putting strong signals into receivers which are tuned to local stations having frequencies close to or the same as those of the interfering stations.

There is nothing that can be done about this by the listener: it is just the result of the crazy congestion which has been developing steadily over the years and must. unfortunately. be accepted by those whose desire it is to tune to the top end,
in particular, of the medium-wave band after dusk.
There is no need for listeners of the local programmes to suffer the disturbances, however, for the v.h.f. service, Jaunched at very great expense by the BBC for the purpose of improving local reception, can solve the problem.

As is shown in Fig. 1(a) skyward signals from a v.h.f. transmitter normally fail to be reflected by the layers in the upper atmosphere, which means that distant stations cannot normally affect local reception, even though frequency sharing may be adopted.

## Copture Effect

There are times. however, when the v.h.f. signals are reflected due to abnormal weather conditions and during settled periods during the spring and summer months. The f.m. system can handle this problem, nevertheless, and this it does because of an f.m. characteristic-called the "capture effect"
-which causes the f.m. set to lock on to the strongest signal and almost completely reject the weaker interfering signal, even though this may be on the same frequency as the wanted signal.
Thus, listeners to the f.m. service are pretty well assured of good interference-free reception under all conditions. That is, provided their receiving installations are correotly tuned and adjusted.

One of the reasons why good v.h.f. reception is not achieved in practice in some cases is that the signal applied to the set from the aerial is far too weak. Nearly all commercial f.m. and a.m./f.m. receivers contain an internal compressed dipole for Band II operation, and while this may deliver sufficient signal when the set is operated reasonably close to a transmitter and in a building and local area which is not unduly screened, it is far from adequate at distances in excess of about 10 miles from the station.

Modern sets are remarkably sensitive, and even on a very weak signal they will give some sort of results, and quite loud signals at full volume, but operated under this condition any local interference will almost certainly be stronger than the signal itself.
Unless the set (or tuner) is a very elaborate type, the interference will break through as crackles on sound, and this applies especially to car ignition and electrical interference from domestic appliances.

## Amplitude Limiting

Given a reasonably high level of wanted signal, all but the very poorly designed and inexpensive receivers will automatically suppress normal electrical interference due to the amplitude limiting action which is inherent to f.m. sets.

All interference of this kind is amplitude modulated. Thus, it follows that a set which is sensitive only to f.m. signals and considerably insensitive to a.m. signals will give interference-free results even when operated in an interference-infested area that would render ordinary a.m. reception almost impossible.


Assuming, then, that the f.m. set is in good order and correctly tuned (for incorrect tuning will affect the pick-up of interference, as we shall see later), a high level of interference should lead immediately to a check of the aerial system.

If nothing more than the internal aerial is used; one could be fairly safe to expect a "dramatic" improvement in reception by the use of an external aerial. Unless one happens to reside in an area outside the range of an f.m. station (see Table 1), the external aerial need not be too elaborate.

Often a simple room or loft type aerial is sufficient, provided it is cut to the optimum length, turned for maximum signal pick-up and connected to the receiver through the type of cable (twin feeder or coaxial) as recommended by the manufacturer.

## Basic F.M. Aerial

Fig. 2(a) shows the basic f.m. aerial. This consists of a dipole, comprising two lengths of conductor each of dimension 1, giving an overall length of 2 times 1 plus the length of the insulator. In practice, the distance between the holes in the insulator, carrying the two halves of the dipole, is about lin., but this is not critical.

For all areas in Great Britain, the overall length of an f.m. dipole is 5 ft . $2 \frac{1}{2} \mathrm{in}$. Thus, the length of each half will depend upon the length of the insulator employed.

To the centre of the two conductors is connected the feeder cable to the set. The diagram shows the connection to coaxial cable, but the connection to twin feeder would be similar; that is, one conductor to each half of the dipole.

If screened twin feeder is used, the outside screening braid should be dressed back and left disconnected.

Now, an aerial of this nature, when mounted horizontally, as is required by the horizontally polarised signal from the f.m. stations, is bidirectional. It has two directions of maximum response, and two directions of minimum response, as shown in Fig. 2(b).

For the best results, then, the aerial should be positioned so that it looks broadside on towards the station. If necessary, a map should be used to establish the direction of the station relative to the receiving site.

Two practical aerials which can easily be constructed by the enthusiast are shown in Fig. 3. At (a) the dipole is formed by two lengths of insulated wire supported between a pair of insulated hooks. in the roof-space, while at (b) the dipole is made of $\frac{1}{2}$ in. diameter tube supported on a wooden frame and mounted on insulators. Both types will give good reception up to distances of about 25 miles from a station.
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Fig. 3: Practical aerials. (a) Roof-space type. (b) Selfsupporting on wooden frame.


When installing the aerial in the roof-space it is desitable to keep it as clear as possible from metal pipes and water tanks, as these can affect the pickup efficiency and tend to encourage instability rat the set.

## Multipath Interference

A cause of poor f.m. reception not generally known is multipath interference due to the v.h.f. signal being reflected from a nearby hill or large building and arriving at the receiving aerial a very small fraction of a second after the arrival of the direct signal. Readers versed in the arts of television reception will know that such reflections of the television signals produce ghost pictures on the screen.

The direct and reflected signals at the f.m. aerial produce a characteristic distortion which, as already intimated, sounds not unlike a badly adjusted loudspeaker. Should this trouble be experienced, a check should first be made to ensure that the set itself is free from blame, and the best way of doing that is either to run it on records, if a radiogram, or to switch to a.m. reception if of the a.m./f.m. variety. If the effect is present on an f.m.-only set or f.m. tuner, it may pay to try it out at a different location, on a friend's aerial, for example.

However, the cure for the multipath interference
is cutting out the reflected signal or, at least, making the strength of the direct signal about 100 times that of the reflected signal. This discrimination can only be handled at the aerial-there being no way of tackling the problem at the set itself.
With a directional aerial, the orientation can be arranged so that the reflected signal arrives along the axis of minimum response, assuming, in the case of a simple dipole, that the direct signal is arriving approximately at right-angles to this.

## Multi-element Aerials

In severe cases of multipath interference, an aerial of greater directivity than given by the simple horizontal dipole may be required. This would comprise the dipole and a director-or a dipole, director and reflector (or a dipole and reflector). The director and reflector are called "parasitic elements" because they are not connected electrically to the dipole.

Their purpose is to strengthen the pick-up of the signal along the axis of maximum response and further to reduce the pick-up along the axis of minimum response. What happens is that the aerial becomes responsive to signals arriving essentially along one major axis, as shown in Fig. 4.


Fig. 4: The addition of a director and reflector increases the forward pick-up and reduces the pick-up at the rear. This type of aerial may be required where multipath interference is troublesome. (Dimensions in text.)

This improved directivity makes it more feasible for the aerial to discriminate between the direct and reflected signals, especially if the reflected signals are arriving at an angle close to that of the direct signal.

Note from Fig. 4 that the director is placed in fromt of the dipole and the reflector at the rear. The reflector is always a little longer than the dipole and the director a little shorter, optimum lengths being 5 ft . 6 in . and $4 \mathrm{ft} .10 \frac{1}{2} \mathrm{in}$. respectively. The spacings are also critical, being 32 fin . from reflector to dipole and $19 \frac{1}{2}$ in. from dipole to director.
Distortion due to multipath interference can often be reduced simply by altering the direction of the aerial. Sometimes this will result in a decrease in pick-up of the direct signal, but provided the signal is still sufficiently strong to overcome car and electrical interference this does not matter much.

However, if the direct signal is greatly reduced at the point of minimum distortion, a small transistorised booster could be used to restore the level of the signal.

## Signal Strength Indication

Most tuners and receivers have some sort of tuning indicator for f.m., and this is useful for giving a rough idea as to how the signal pick-up has been increased or decreased by orientation of the aerial, by the use of a booster and by the use of different aerials.
Not only multipath interference, but interferences of other types are aggravated by the use of an incorrect aerial. For example, one may be tempted to use the television aerial on f.m. This invariably works. but optimum results are obtained only if the aerial system has Band II facilities. Television sets equipped with f.m. are often expected to work from the television aerial without an f.m. attachment.
This compromise is bad because the television aerial is not cut to the lengths required for Band II reception and also because the f.m. signal is horizontally polarised (needing a horizontally mounted aerial) while the TV signal in many areas is vertically polarised (with the aerial mounted as a vertical "H" for instance).
It then often happens that at times when interference from distant stations is troublesome, as explained earlier, the interfering signal picked up on the incorrect aerial is stronger than the wanted local signal! This causes the "warbling" interference effect on top of one (or more) of the local f.m. programmes.

## Combined Aerials

Fig. 5(a) shows how an f.m. aerial can be connected along with the TV aerials, so that all the signals pass through the common coaxial downlead. If it is required to separate the signals at the set end of the downlead, the arrangement shown in Fig. 5(b) should be employed. At (a) the signals are combined by a triplexer while at (b) they are separated by an f.m. diplexer (passing the TV signals at one outlet and the f.m. signals at the other).
Where installation is on a main road and passing traffic interference is troublesome in spite of a good aerial system, use of one of the small transistorised f.m. boosters can be of great assistance. This is because the greater the signal applied to many sets, the better the amplitude limiting.

Thus, if the signal without the boost is just about strong enough to commence the limiting action, heavy car interference will cause trouble. The extra boost, however, often pushes the set well into limiting and considerably reduces the interference effect.

A little extra attention should also be given to the tuning-in of an f.m. set, for unless the tuning is accurate the full attributes of the f.m. system will not be realised, and both distortion and interference could cause trouble. There are usually three tuning points close to where the station is tuned, two small responses either side of the main, maximum response.

It is imperative that the set is tuned to the


Fig. 5: A triplexer is used at the top of the downlead to connect an f.m. aerial (a). An f.m. diplexer, or similar device, must be used at the bottom of the lead if it is required to separate the signals, as shown at (b).
maximum response. Always tune for maximum deflection as shown on the tuning indicator, where fitted.
If there seems to be two major tuning points close together, the set or tuner itself may be out of alignment, and this, again, is a fundamental cause of poor f.m. reception. A f.m. set is not easily aligned properly without instruments and since so much relies upon correct alignment, it often pays to have a good radio dealer perform the necessary trimmer adjustments.
Since the many f.m. stations are placed with both geographic and frequency consideration, interference from adjacent channel signals does not occur under normal conditions. This means that full expansion can be given to the sidebands of the higher audio frequencies, and treble cut and passband cutting, which are essential on medium frequencies, are not required on f.m.
Thus, the quality of reception on a correctly working installation is far above that of the a.m. system. However, incorrect alignment can suppress the treble and make tuning difficult. Moreover, if the i.f. passband is restricted, excessive tuning drift will be noticed as the set warms up and as its temperature rises during use.

## Tune to the Local Station

One final point, and that is since pretty well every local area has its own f.m. transmitter now, it is possible to pick up the signals. from transmitters in adjacent areas. These signals, being weaker than the local signals (although carrying the same programmes), will be more prone to interference.
One should always ensure, therefore, that the receiver is, in fact, tuned to the local transmitter and not to an adjacent area one.

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An up-to-date introduction to electronics which assumes no prior technical knowledge of the subject. Early chapters explain the nature of electric currents, pulses and waveforms. The components and circuits that are the basis of electronics are then described and illustrated, and their operation clearly explained. Chapters are devoted to test instruments; the principles of and the basic techniques used in the main branches of electron-ics-radar, radio, television, medical electronics, electronics in space, industrial electronics and computers, and on training to be an electronics engineer.

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# IMPROVED COMMUNICATIONS THROUGH THE GROUND 

Same further notes on thos novel method of transmission and reception previously described in P.W.

BY<br>C. R. BR.ADLEY

|N the May 1964 issue of Practical Wirfiess, an article of mine was published on a system of communication through the ground. I was quite unprepared for the tremendous response I received to this article and wish to thank all readers who wrote to me-if replies have been delayed it does not mean your letters have not been appreciated.

Well over a hundred letters arrived in the first week and they are still coming in from abroad! Most of the letters have been from readers who have had success with the method-a Rew thought
the whole thing was a hoax!
I have received many practical results and much historical information from readers on the subject. some of which I take the liberty of reproducing here.

The kind of ground communication described in my article was used extensively during the 1914 1918 war for morse communication through heavily shelled areas. The equipment was bulky and often in poor repair, yet ranges of up to 8,000 yards were achieved!

It seems however, from readers' letters, that the greatest distance practical with hi-fi or guitar amplifiers is about two miles. Results with amplifiers with power ratings below 8 W have been disappointing; this is not surprising considering the tiny fraction of the transmitted signal which reaches the receiver.

The resistance from point to point of the ground varies so much with locality that no convenient formula giving the maximum achievable distance for a particular transmitter power can be found. The strength of the received signal will depend on a multitude of factors, including soil moisture, temperature and type.

Generally however, best performance is obtained by having both transmitter and receiver earths as far apart as possible and set in roughly the positions shown in Fig. 1b.


Fig. 2: A multi-directional earth system. The two-pole three-way switch selects different palrs of earths and hence different directions.

A certain amount of mains hum is always received: this is caused by a.c. currents flowing through the ground, due mainly to the earthing of mains equipment. This hum limits the weakest signal that can be received but is only prominent iti built-up areas. There is no way of reducing it.

Some readers have encountered interference from nearby radio stations and even wired relay systems. In one case, ground communication experiments were found to interfere with a neighbour's telephone-unfortunately no cure could be found and they were terminated.

Because of the limited (but useful) range of ground communication there seems no need for a national "Ground Ham" club although it may prove an interesting sideline for existing local radio clubs. I am pleased my article has stirred up so much enthusiasm for a long forgotten idea which shows how far from "dead" the ground can be!

## A COMMENTARY BY HENRY practicall WIRELESS

## No. 6 <br> Trouble Aloft

THE Fill-Dyke month: when gutters are full of sleet, gales howl about our ears, and Henry feels as old as his children imagine, him. The time, in short, when the aerial blows down.

Aerials-yes, those remote and spindly refugees from a mordern art show that so many radio users still believe they are simply an adjanct to television. Aerialsthe first, most important, and most of ten reglected link in the chain of radio reception. Many of us continue to believe that a few microvolts in the hand are worth any amount of watts in the bush. Consequently, our eaves groan in the gales, and the ice on the insulators makes a climb up the mast a daunting prospect.

Which reminds me of a story, which really I should not tell you, for it concerns a television aerial-and my spies tell me there is another magazine across the way which specialises in that upstart offshoot.

The adventure happened in the very early days, when there was only one Dimbleby and folks used to tune to the morning test signal for the fun of watching the Palio horse-race. I was called to investigate a set with no modulation, and immediately jumped to


An enthusiastic teenager and a hint of the Big Beat.
the obvious. Sure thing, a dead short across the aerial.
Defying the gales and risking everything on the sleety tiles, I shinned up to the chimney and began to unship what seemed like half the GPO tower. It was not until the ladder was shaking beneath my descending clodhoppers that I realised what I was carrying. One of the original Duplex aerials, no less. A Band ! folded dipole, which even Henry should have known would give a dead short reading on the ohmeter-if he had only taken the trouble to look.
How many riggers, I wonder, fit a small resistor across the remote terminals of an open dipole for future bad-weather checks on possible open circuits?
It's a bit difficult, however. when one is using the length of the garden for an old-fashioned receptor, and the weight of the snow, aided by the gathering of our feathered population, and augmented by ravages of rust at the staybolts has torn the whole contraption from the garage wall and left a drooping trail of destruction through the cabbages.

Even the dinky. but effective ferrite rod has its drawbacks: mainly the extreme brittleness of the material. Aided by so many manufacturers' designs, where a long, thin rod is supported by a couple of barely padded clamps about a third of the way allong.

Given an enthusiastic teenager and a hint of the Big Beat and the end of the rod droops dismally, like the end of a comedian's cigar. Even where the rod is supported more adequately, its dimensions and its brittleness make it vulnerable to shock.

Where so many folk. get caught, we regret, is in writing the broken rod off as a dead loss. this device is a magnetic, not an electrical, circuit. Thus, we need only reassemble the parts, particularly if the break was clean, to regain the flux concentration.

Provided the rod can be put


Some wicked manufacturer, with a warehouse full of replacement rods.
together really well, with split ends butting. splinted like a C.D. experiment and, tightly bound, there should be very little loss in aerial gain. Such loss as is apparent can often be overcome by slightly repositioning the coils. As any reader with a loose coil will have found, the positioning is quite critical.

One day, we presume, the boffins will come up with a material less prone to stress failure. They may have done so already. for all 1 know. Maybe some wicked manufacturer. with a warehouse full of replacement rods, is sitting on the patent rights.

In my pocket I have a useful aid to tape recordings, a piece of magnetic rubber. This is what it says it is - rubber with magnetic properties, made by impregnating the rubber in the forming stage with tiny magnetic particles. As a similar process is used in the manufacture of ferrite rods. but with ceramic instead of rubber as the base. perhaps some genius will think of a way of making them more flexible
In the meantime. I must return to my frozen-fingered efforts to splice steel wire through the egg insulator before the 20 metre band re-opens.
End of Febrifuge-over and out.

## Pioneers in build it yourself tuners and test equipment

JTV2 SWITCHED TUNER for the reception of all BBC f.m. transmisslons and BBC-1 and ITV television sound channels at the turn of a switch. Foster-Seeley discriminator. Built-in power supply. Sensitivity $10 \mu \mathrm{~V}$ for 40 dB quieting. Audio output 0.4 V approx. All components including turret and coil plate but lass four valves 614.0 .4 .


FMT2 F.M. TUNER covering $88-108 \mathrm{Mc}$, . Euilt-in power supply. All components, less four valves 69.9 .0 , less power supply components $\mathbb{\text { f }}$. 15.0 .
FMT3 TUNER similar appearance to above, sur for fringe area reception. All comporents. less valves $\mathbb{C} 10.1 .0$, less power'suppily components $\mathbf{E 8 . 1 5 . 0}$.

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## TRADE NEWS•TRADE NEWS • TRADE NEWS TRADE NEWS•TRADE NEWS • TRADE NEWS

## New Sin. Oscilloscope Tube

,1ULLARD announce the D13-27GH, a further addition to their range of oscilloscope c.r.t.s.
The new D13-27GH is an inexpensive sin, tube with an overall length under $14 i n$. and this compactness enables it to be substituted with advantage for the small screen types often employed in monitor applications.

This tube has a flat faced, medium persistence green phosphor screen with a helical post deffection accelerator. In common with other recently introduced Mullard tubes, a separate electrode arrangement permits direct beam blanking to be accomplished. The deflection blanking voltage required to blank the tube is 60 V maximum under 3 kV operating conditions.

The deflection sensitivity, also under 3 kV operating conditions, is better than $27 \mathrm{~V} / \mathrm{cm}$ for the X direction and $13 \mathrm{~V} / \mathrm{cm}$ for the Y direction. Mullard Lid., Mullard House, Torrington Place, London, W.C.I.

W.K. \& C. Peace's handy tool kit in plastic roll-pock wallet.


Model 450 valve tester from Toylor Electrical Instruments Ltd.

New Valve Tester
TAYLOR Electrical Instruments (M. I. Group) has introduced a new modern styled valve tester, Model 45D. This supersedes the Model 45C.

Main features of the 45 D are the ten valve bases which enable tests to be carried out on the latest types of valves.

A valve chart which is included gives testing data for over 7.000 British, American, Continental and Russian valves. The manufacturers of this instrument are Tavlor Electrical Instruments, Muntrose Avenue, Slough, Buckinghamshire.

## Amateur Radio Equipment

NEW additions to the Ancerican Heathkit range of amateur radio equipment are: single sidehand receiver, model SB-300E at $£ 133$ 14s., single sideband transmitter SB-400E at $£ 1654 \mathrm{~s}$. and 80 m s.s.b. transceiver HW-12 at $£ 60$ 1s., and a single monitor oscilloscope HO-10E priced at $£ 3410$ s.

Instruments in the Heathkit range that have been re-styled are the 5 in . oscilloscope $10-12 \mathrm{U}$ for $£ 32$ 125. 6d., television alignment generator HFW-1 at $£ 34 \quad 18 \mathrm{~s}$. . and the 6 in. valve voltmeter $1 \mathrm{M}-13 \mathrm{U}$ priced at 18 guineas. Daystrom Limited. Gloucester, England.

## Tope Splicer Kit

MASTERTAPE announce the introduction of a new and improved Mastertape splicer kit retailing at 21 s . and containing a tape splicer and cutter, five different coloured leader tapes, metallic strip and jointing tape. Full operating instructions for the splicer are printed inside the lid of the distinctive carton. MSS Recording Companv Limited, Poyle Trading Estate, Colinbrook, Slough Buckinghamshire.


Baird's model 100 transistorised stereo radiogram.

## Stereo Radiogram

MODEL 100 is Baird's first stereophonic radiogram. It is completely transistorised and power is obtained via a mains transformer so isolating the chassis.

Separate bass and treble controls are incorporated and there is a balance control for equalising the volume of the two audio outputs. Waveband selection is by push-button control.

There is provision for connecting extension speakers separately to each audio channel so that, where required, full stereo reproduction is available at an extension position. A tape recorder socket is incorporated to provide facilities for recording from radio or records, and tape recordings can be

## PREPARING FOR THE R.A.E.

-continued from page 973
resonant-the secondary circuit being a link winding. In $C$ the secondary circuit only is resonant-the primary being a link winding and in (d) both circuits are resonant. (a), (b), (c) and (d) couple the circuits in different ways but (b), (c) and (d) are all coupled because of the mutual inductance between the coils.

### 4.9 Acceptor and Rejector Circuits


(a)

(b)

Fig. 36-Acceptor and rejector circuits.

A series connected circuit, as shown in Fig. .36a is called an acceptor circuit, because if two equal voltages, at different frequencies, one of which was the same as the resonant frequency of the circuit, were applied to the circuit the maximum current would flow for that voltage at the resonant frequency, i.e. the resonant frequency is more readily accepted.

Conversely in Fig. 36b the circuit is a rejector circuit, because at the resonant frequency of the circuit the current in the circuit is minimum.
replayed through the stereogram amplifier/ speaker system.

Built-in aerials are fitted for both a.m. and f.m. and there is provision for rotating the ferrite a.m. aerial. There is also provision for the connection of an external f.m. aerial.

The cabinet is finished in high gloss polyester and adjustable glides are fitted to the legs.

The circuit embodies 19 transistors and 4 crystal diodes. The record changer is a B.S.R. UA15 fitted with a B.S.R. stereo dual stylus.

The overall size of the Model 100 is 504 in . wide, 28 in . high and $14 \frac{1}{2} \mathrm{in}$. deep, and the price is 70 guineas including purchase tax. Baird TV Distributors, Lid., Empire House, 414 High Road, Chiswick, London, W.4.

## Morse Records

RECENTLY released by The Morse Centre is
"The Beginners' Course", a new addition to their range of morse records. The Beginners' Course consists of two records, one 12 in . instruction record plus one 7 in . EP containing two simulated GPO amateur morse tests. Each is available at 9,12 and 21 words per minute, plus an instruction book. The cost of this course is $£ 3.0 \mathrm{~s} .6 \mathrm{~d}$.

Also available is the Complete Course which consists of three records as well as instruction books, a Beginners' ${ }^{\prime}$ in. LP, an Advance Students' 12 in . LP and a 7 in . EP, all of which are played at three speeds. The overall cost, including postage in the UK is $£ 44 \mathrm{~s}$. The Morse Centre, 45 Green Lane, Purley, Surrey.

## Question

In Fig. 37 find at what frequency the circuit would be resonant and calculate the value of the capacitance


Fig. 37-Inductance and capacitance in parallel across an a.c. supply.
which would have to be placed in parallel with $C$ to make the circuit resonate at half this frequency.

## Answers to last Month's Problem

(a)

$$
\begin{aligned}
& \begin{aligned}
& \mathrm{X}_{\mathrm{c}}=\frac{1}{2 \pi \mathrm{fc}}=\frac{1}{6 \cdot 28 \times 50 \times 0.000002} \\
&=\frac{1}{0.000628}=1592 \Omega \\
& \mathrm{E} \quad \mathrm{E} \\
& \mathrm{I}=\frac{250}{\mathrm{R}} \quad=\frac{15}{1592} \\
&=0 \cdot 157 \mathrm{amps} . \\
& \mathrm{X}=2 \pi \mathrm{fL} \\
& 1592=6.28 \times 50 \times \mathrm{L} \\
& 1592=344 \cdot 00 \quad \mathrm{~L} \\
& 1592
\end{aligned} \\
& \therefore \mathrm{~L}=\frac{1}{344} \\
& =\mathrm{L} 4 \cdot 63 \\
& \text { Value of Inductance is 4.63. Henrys. } \\
& \text { Part } 5 \text { Next Month }
\end{aligned}
$$



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& 200
\end{aligned}\right.
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by D. Gibson, G3JDG



IT is sometimes said that an amateur station can be judged by its auxiliary equipment. The station operator who has transmitter, receiver. and frequency meter as the sum total of his possessions can barely claim to be a "ham".

The subsidiary equipinent mostly consists of small easily built units and any excuses such as lack of time or ready cash are, to say the least, a little thin. A grid dip meter is no more than a single valve oscillator with its grid current metered. An absorption wave meter consists of even less.

When it comes to loading the aerial and radia. ting r.f. the need for a means of tuning and loading accurately is definitely desirable. At this point, our tyro with minimum equipment will triumphantly exclaim that his pi-tank matches most things he feeds it into. That the meter situated in the h.t. to the p.a. gives all the indication that is needed as regards loading, and that by using the "tuning for a slight dip" technique, contacts have been made.

Since the proof of any pudding is in the eating, it should, therefore, surely follow that the p.a. is working efficiently. loading is correct and that the radiated r.f. furnishes all the necessary proof of this. Unfortunately, tuning for the proverbial slight dip does not mean that everything is working efficiently.

It is possible to tune a rig by this method and have a standing wave ratio much higher than is either desirable or necessary, and the higher the s.w.r. on the line, the less efficient the whole thing becomes. It is not proposed to split technical hairs over this business of s.w.r., nor is it proposed to launch into a discussion on transmission lines, as books exist which already do this.

However; whenever experts meet. the first thing they usually do is to define terms to make sure that they all mean the same thins: and that they are all "talking the same language". What is good enough for the experts is good enough for us.

When we say that a standing wave ratio exists we shall take it to mean that some of the power which should have radiated into space from the aerial is, in fact, being refiected back down the feeder, and thus the efficiency of the aerial system
is not $100 \%$. In other words, we are wasting power, heating the coaxial, or warming the worms.

Obviously, a device which can give a direct indication of this s.w.r. is very desirable because it would allow things to be adjusted until a minimum was obtained. Fig. 1 shows a simple arrangement used by many amateurs and SWL's. It is a halfwave dipole, and is fed at the centre with 7512 co-ax, thus offering a good match.

The centre of the dipole, has an impedance of around 7512: the coaxial has an impedance of 759 so any r.f. proceeding up the coaxial "sees" no difference of impedance, and we have a perfect match. (For the purpose of simplifying matters we are ignoring the fact that the dipole is a balanced system and that coaxial is not.) The s.w.r. is said to be $1: 1$.

Our dipole is only a perfect match, however, at one frequency, the frequency for which it is cut, and directly the rig is tuned up the band a few $\mathrm{kc} / \mathrm{s}$ we have a mis-match. The dipole no longer exhibits a pure resistance of $75 \Omega$. It now shows reactance as well. and this reactance will either be cuapacitive or inductive depending on whether we are ahove or below the original frequency. All this boils down to the fact that now, we have a s.w.r. on the line.
NOTE:-Tuning our pi-tank for the above mentioned dip will not tell us much. As far as the meter is concerned it merely indicates a dip in p.a. current. An r.f. ammeter in the feeder is of little


Fig. 1: A simple half-wave dipole.


Fig. 2: This arrangement although favoured by some does not, however, provide much information about the s.w.r.
use in informing us of any s.w.r. and its reading would vary depending on where it was situated in the feed line.

This, of course, applies to odd lengths of wire tuned up by our faithful pi-tank on different bands. On one band it may be at a point in the system of high current and indicate a reading of say " 7 " and on another band it may be at a point of high voltage where very little current exists, and the maximum reading on this band may well be only 1 or 2.


Fig. 3: The complete circuit of the instrument, representing a considerable improvement on Fig. 2.



Fig. 4: Stages in the construction of the coaxial + conductor element of the circuit.

One possibility is to construct an r.f. voltmeter to pick up a "sniff" of r.f. off the line, rectify it, and apply the resultant voltage to a suitable meter. This circuit can be very simple, and also very small in size (Fig. 2). This method is favoured by some, but once again although it now indicates output, and not just p.a. current it still tells us very little about the s.w.r.

If we are still determined to have some means of visual indication of s.w.r. we may well delve into the appropriate text books and find that a circuit exists which will perform this elusive function. It is referred to by the very grand name of an "H.F. Reflectometer".

The arrangement seen by the writer, required that "the concentric line be converted to a parallel plate line ". It then went on to describe how the various strips of the conductors were mounted with pieces of perspex spacers and lengths of $16 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. wire. The setting up of the monster so frightened the writer that it was thought that perhaps our tyro has .a point. So long as sufficient r.f. was radiated to bring contacts why worry?

It was not until a circuit appeared which was a brilliant piece of simplicity called the Moni Match that things really appeared interesting. This circuit suggested that instead of two diodes and attendant "Meccano" of the concentric/parallel plate line business, why not use an ordinary piece of coaxial and add to it a length of ordinary thin enamelled copper wire and switch the r.f.? Result-super simplicity, compactness, and a reduction in the number of components required.

The circuit of the entire instrument is shown in Fig. 3. Note that it is not unlike Fig. 2 but with a very great advantage that it can read the output or reffected wave (s.w.r.) merely by flicking a switch. It -uses components, which are easily obtainable and inexpensive, and the entire instrument might perhaps be constructed in one of the sloping fronted meter cases available.

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## ALPHA RADIO SUPPLY CO., 103 leeds terrace, wintoun st., LEEDS 7




The unit shown was constructed in an aluminium case because an a.t.t. was also built into the same case. The a.t.u. is nothing more than a coil and capacitor both purloined from one of the T.U. series. The meter used is a $0-500 \mu \mathrm{~A}$ and is from the ex-Govt. 19 set. These are freely available at 155. or less on the market, although any similar meter would be suitable.
The only part of the unit requiring any special treatment is the length of the coaxial with its added conductor, and this is really quite a simple matter when it comes down to actual construction.
Take a piece of $75 \Omega$ coaxial as shown in Fig. 4 and bunch up the braiding a little. Next take one of the XYL's sewing, needles and thread a length of 34s.w.g. enamelled wire. Push the needle under the co-ax braiding about $\frac{1}{2}$ in. from the end and work the needle along and out of the coaxial $\frac{1}{2}$ in. from the other end (By means of a magnet perhaps?).
Warning:-Work the needle along, don't work the wire, you might accidentally scratch off some of the insulating enamel.
Next, check with the meter that there is no short between the piece of enamelled wire and the coaxial braiding. A small scrap of p.v.c. tape around each end completes the job. Apart from keeping all leads (especially around the switch) as short as possible the wiring-up requires no further comment.

When the unit is completed, it has to be calibrated. and. to do this a dummy load is required. A carbon resistor of the same value as the line is required i.e. for $75 \Omega$ coaxial a $75 \Omega$ resistor is required. It should be capable of withstanding the power output of the transmitter. and a number of resistors may be wired in series/parallel if desired or the transmitter power may be reduced.

For those who are unable to obtain a $500 \Omega$ carbon pot. (R1), a number of carbon resistors may be substituted one at a time until a good null has been obtained (See notes on setting up). Although this method may take a little longer it is an improvement on using a pot.. since any slight reactance offered by the pot. is then eliminated.

Fig. 5: The front panel of the controls, sockets, etc. SK2 could take the alternative position indicated.


## Setting-up

The actual setting up follows a logical pattern, and if taken a step at a time the whole process is fairly straightforward.

1. Plug transmitter output into SK1 and connect dummy load to SK2.
2. Switch SI to forward and tune transmitter. This is best done on the highest band to be used.
3. Adjust R2 so that M1 reads f.s.d. or highest reading on its scale, which ever is the greater.
4. Turn $S 1$ to reflected and again adjust $R 2$ for maximum reading or f.s.d. whichever is the greater.
5. Adjust R1 for minimum reading.
6. Switch back to forward and readjust R2 slightly for maximum reading on M1.
7. Switch S1 to reflected and check that the null is still there.
8. Swap the transmitter and dummy load i.e. plug transmitter into SK2 and dummy load into SK1.
9. The readings should still tally near enough with the previous ones obtained in 6 and 7.
10. The unit is now ready for use, the transmitter is plugged back into SK1 and the dipole or a.t.u. etc., into SK2.


## BRADFORD RADIO SOCIETY

Hon. Sec.: Eric G. Barker, G3OTO, 63 Woodcot Avenue, Baildon, Nr. Shipley, Yorkshire.
'On 15 th December there was a Ladies' Night, and members brought along their colour slides for a film show.

At an informal meeting held on 5th January, the guests were members of the Spen Valley Amateur Radio Society and the Leeds Radio society.

## BROMSGROVE AMATEUR RADIO CLUB

J Harvey, 22 Elm Grove, Bromegrove, Worcesterwhire.
A meeting will be held in the Co-operative Rooms (corner of High Street and Stratford Road, Bromsgrove) on Friday, 8th January, at 8 p.m.
As this was an inaugural meeting, radio amateurs and short wave listeners wore invited to attend with suggestions and comments.
CHESTER AND DISTRICT AMATEUR RADIO SOCIETY Hon. Sec.: P. J. Holland, Field House, 19 Kingsley Road, Great Boushton, Cheter.
A technical film show was seen by mermbers on 8th December, and on the 1Sth there was a Christmas Surprise Night. On 22nd Decomber there was a general discussion and the month's activities concluded with a Net Night on 29th December.

## CLIFTON AMATEUR RADIO SOCIETY

Hon. Sec.: J. Rose, G3OGE,63 Broomfield Road, Beckenham. Kont.
The innual Constructional Contest was held on th December, and an informal social was held later on in the month.
Workshop improvements have now been completed, thanks to a handful of members, and meetings are held on Wednesday and Friday evenings.

## COVENTRY AMATEUR RADIO SOCIETY

Hon. Sec.: E. E. Snow, G3TKO, II Lupton Avenue, Coventry. On 7th December there was a film show, and on the 14th there mes a demonstration and Ifecture on construction of the Club top band transmitter. On 21 st December there was a social evening, and on the 28th, A Night on the Air. A film show was seen by members and friends on 4th January.
DERBY AND DISTRICT AMATEUR RADIO SOCIETY Hon. Sec.: F. C, Ward, G2CVV, 5 Uplands Avenue, Littleever, Derby.
There was a surplus sale on $2 n d$ December, and on the 9 th there was a constructors' contest for the Founder Members' Trophy, with e contest for the G5YY Trophy on the 13th.

The I6th December was an Open Evening with a Juniors' meeting and members committee meeting. The Annual Christmas Party was held on 23 rd December and on the 30th there was a film show entitled "The Year in Retrospect" when members brought along transparencies which they had taken during the year.

## GOSPORT AND DISTRICT RADIO CLUB

Hon. Sec: J. T. Nightingale, 9 Beatty Drive, Alverstoke, Gosport.

This Club's meetings are held every Monday evening at 7 p.m. the Gosport Community Centre, Bury House, Bury Road, Gosport. Local radio enthusiases and would-be enthusiasts are walcome to go along and join in the numerous activitios.

## GUILDFORD AND DISTRICT RADIO SOCIETY

Hon. Sec.: D. H. Mead, G30xi, 41 Egley Road, Woking, Surrey.

The speaker on Ilth December was Mr. M. Childs, F.S.A., President, who spoke on "The Operation of Practical Equipment from $1900-1922^{\prime \prime}$.

On 19th December there was a Christmas Social Evening at "The Otter". This was a light-hearted evening in the seasonal spirit to which all members, their ladies and friends ware invited.
HALIFAX AND DISTRICT AMATEUR RADIO SOCIETY Hen. Sac.: J. Ingram, G3RM a, Lambert House, Greetland, Helifax, Yorkshire.
On 2\%th December, members brought along their receivers and heid a general discussion on their equipment.
On 26th January, there will be a talk entitled "Radio a long time Igo", given by Mr. J. J. Platt, G2VO.




HUDDERSFIELD AMATEUR RADIO CLUB
R. Higton, 5 Brian Avenue, Dalton, Hudderefield, Yorkshire.

Meetings are held fortnightly on Thursday evenings, beginning at 7 p.m. The evening usually begins with $\mathrm{c} . \mathrm{w}$. practlce for the short wave listeners, followed by lectures. Field days and outings are held regularly. Members were very pleased with GB3CMS which was the "Jamboree on the Air" station.
MELTON MOWBRAY AMATEUR RADIO SOCIETY
Hon. Sec.: D. W. Lilley, G3FDF, 23 Melton Road, Ashfordby Hill, Melton Mowbray, Leicestershire.
On 3lst December, Mr. L. Fisher, G4MK, and Mr. D. Fisher, gave a talk on the use of test gear. There will be a lecture on 2beh January, by Mr. J. L. Warrinston, G2FNW, entitled "Flat Line Equipment for 23 cms .'
MID-WARWICKSHIRE AMATEUR RADIO SOCIETY Hon. Sec: T. Inkester, 13 Dormer Place, Leamington Spa, Warwickshire.
The 2nd November maeting saw the inauguration of a series of informal lectures on the fundamentals of radio. These lectures will alternate with evenings at which the newly acquired A/T gear will be operated by G3HBX/A.
NORTHERN HEIGHTS AMATEUR RADIO SOCIETY
Hon. Sec.: A. Robinson, G3MDW, Gandy Cabin, Ofden, Halifax ${ }^{\text {M }}$ Yorkshire.
The Club's Annual Dinner was held on 9th December, and a ragchew was held on the 23 rd .
On 6th January A. W. Walmsley, G3ADQ, gave a telk ontitled "SS8 Trends"
PLYMOUTH RADIO CLUB
Hon. Sec.: R. Hooper, G3SCW, 2 Cheatnut Road, Paverell, Plymouth, Devon.

On Saturday, 19th December, the Club supported the Torbay Amateur Radio Society Christmas Party and Quiz.

Saturday, 6th February, is the Club's Annual Dinner and Social Evening.
READING AMATEUR RADIO SOCIETY
Hon. Sec.: R. G. Nash, G3EJA, "Peacehaven," \& Holjbrook Road, Reading.

The meating hald on 19th December wes dovoted to the judzing of the Constructional Contests.

The Annual Dinner will be held at the White Hart, St. Marr's Butts, Reading, on 9th January.
SLADE RADIO SOCIETY
Hon, Sec.: D. T. Wilson, 177 Dower Road, Four Oaki, Sutten Coldfield, Warwickshire.

An annual Fun and Games Evening was hold on IIth Decambar. The evening included the annual Cricket Match between the President's and Secretary's teams.
WELLINGBOROUGH RADIO CLUB
Hon. Sec.: J. Baker, 34 Essex Road, Rushden, Northemptonshire.

On 3rd December, members heard a talk on Radio Control. There was a General Meeting on the loth and the annulal Christinas Dinner on the 17 th. On 7th January members taw a film show entitled "The Race for Space".

MR. M. A. TRICKETT, G3SAT, and licensee of club radio $G 3 S O H$ is attempting to obtain a concession in the licence conditions of Club Radio Stations so that anyone may speak into the microphone under the supervision of the licensee or authorised additional operator. provided that the aforesaid actually operates the apparatus.

Mr. Trickett seeks the support of other club members who may be interested in this cause and will forward further information if they will write to him at: 15 Egerton Road, Bembridge, Isle of Wight.

## New list of Bargains

Transiator ferrite rod aerial with medium and long wave coils with circuit. 7/8.
Oscillator Coil and set ot 3 I.F, tranaformers for transistor set witls curcuit, 12/8.
Toning Condenser to sutt. air-spaced with trimulera, $9 /$ -
Ditto but sub, mid. 7nm., 10/- the set; two gang cindensers to suit. $8 / 6$ (request sub. min. circuit)
Midget 3in. P. M. Londapeaket 3 ohm 12/B, 80 obm 13/6.
Midget $208 \mathrm{pF}+17 \mathrm{~A}$ pF two-gang Tuning Condernat
Push-Pall Tranaformer. Bub-miniature 8/8, To05s mid. single Tuning condenser. Solid dielestric tio. spmille for transistor ol 46 Sets (Recelver/Transmitter pack ret). Packed with parta and easily rebuildable tato Packed with pa/s and hear. Pust 3/-
Battery Charger Kit. Comurisera amp. tranelormer, 5 amp. rectither. metal case and meter to charge 6 or 12 volt buttefles 1 p to 5 ampe. and insurance, $3 / 6$.
Mains Transiormer. $250-0 \cdot 250$ at 80 mA . ti.3 vults, ja (norinal mame tnput). $12 / 6$ each.
bartiare $2 / 6$. -sartiage 2/6.
Output Transformer. Standaril pentode matching type. $3 / 6$ each. 38/- por uue. Slide Switeb. Snlo-mlalature but dpdt. $2 f$ each, 18/- per thiz.
T.C.C. of Dubher Tubular Condensers.
$.5 \mathrm{mf} 500^{\mathrm{v}} \mathrm{m}$.

0.5 int 2000
.0001 mil 1,000 v
.001 mf 1.000 v .
$.002 \mathrm{mf} 1,000 \mathrm{r}$.
$.00 \overline{\mathrm{a}} \mathrm{ml} 11.004$
.02 mil 730 c .
10/- doz.

Battery Charger Reotifer-selenum 12..1s v б анир. 9/6.
Metal Chassm-punched for Mullard 510 Amplificr complete witb inner sereening vections and stove enamelled, $12 / 6$ set.
Filament Transiormer, 6.3 v. 1\& ardes.. 6/6. Neon Lamp-midget wire ended. Ideal maina teater, elc. $2 /$-, Fix. Guvt. 1/8.
Phillips Trimmers- $0.30 \mathrm{p} \mathrm{l}^{*} 1 /$ - ea., 9/- doz. Tag Panels. Ideal for constructors, experimental circuita, etc. 3 of each of 12 different tyrea. $5 /$
Slydlok Panel Mounting Fuses with carrier 5 smp. $2 /-$ emch. $\overline{0}$ aull. $2 / 6$ each
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It fuly covers the medium wavehand and that part of the long wave liand to brmg in B.R.C: bights efficent slab aerial and thin. F'. M. speaker. Overall wize ap-


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## THIS MONTH'S SNIP

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sondering frome and looiling ringe up to 2,500 watts 1 umplete aujustable, nommal pricis $55 /$ - each. rpectal svip frice $12 / 6$, plus $1 / 6$ pottare and insuratice

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$80^{\circ} \mathrm{F}, 9 / 8$ plus $1 /$ post, Buitable box for wail mounting, 5/ヶ, P. \& P. 1/Type 'B' 15 anip. This is a 17 in . long rod, trpe made bs the famous sunvig (o spladle adjusta this fram 50 550 F Intermal screw alters the setting so this could we adjustable
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2000 O．F．F．MODRL TP 10 Soe O．F．V．MODRL TP 10 Readn A．C．and D．C．Volte up to Roditance to 1 Mes ，Co plecitance to 1 uF；Deelbela from -20 to +34 ；Output ject for Audio trengrementa flize $81 \times 5 \times$ 1 in $_{\mathrm{cy}}$ 83．19．6．
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 liegure up to bW magahmy at leature up to b0w megahm at Hupplied rith fot eada， 817.10 .0 ． llustrated det －
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 controls．giving a wide range of liff and out geparat mateble gain control．Tremolo apeed and depth controls．Jack bocket for Femote tremolo switching．Outputs for 3 and 15 ohma speakers，Valves used in the 30 watt and 50 watt amplifer ECC83，ECCB3，EL 34, EL34．GZ34．In the 15 watt amplifier LicCe3． ECC83，EL84，EL84，EZ81．An extra valve ECCB3 is used in tho tremolo circuit．The chsssis is completo with baseplite nend is solldy made of 18 gauge steel．finished silver proy hanimit． Size $12 \times 8 \times 6$
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## THE RSGB AND PIRATES

SIR.-I have followed with interest your recent editorials on pirate operation.
I agree there is no excuse for operating a transmitter without a licence-the qualifications should not be beyond the capabilities of anyone who is really keen. In this connection I am particularly pleased to see that you are currently publishing a series of articles designed to help the newcomer pass the Radio Amateur' Examination.

This society has itself been very active in this sphere for many years and publiches in August and Scptember each year extensive lists of centres at which courses in preparation for the R.A.E. are being held. In addition, the society publishes a " Radio Amateurs' Examination Manual". More recently, an education and training committee has been set up, one of the objects of which is to foster the dissemination of information intended to help people to take part in amateur radio to the full.

Yous practical efforts in this field are therefore particularly welcome. - J. A. Rouse (General Manager, Radio Society of Great Britain).

## NATIONAL FRATERNITY

SIR.-We were interested in the letter concerning Boys' Clubs, written by Mr. Overbury in the January issue. Our society and club has been forwarding just such a movement for about four years now. We have had great success in the field of exhibitions and social work. As a society, we have links with boys in Brighton, Welwyn, H.M.S. Ganges (Ipswich) and Australia, and correspondents in India and Nigeria. Work in co-operation with other bodies such as the British Association and. oc course, the R.S.G.B. proceeds internationally together with the "Science Service" of the USA.

However, a club is vital to this work, and forms a focal point for all concerned. Boys in our locality have a club room and equipment, regular meetings, lectures and other club activities. It is vital to set up local activity, and let a larger organisation grow.

Because of the nature of work with young people, their particular outlook, lack of resources and responsibility, recqgition under the Law, the difficulties are. and always have been, peculiar to this type of work.

It is important to have skilled. dedicated leadership. Also, assistance from all those who can give it is absolutely vital.
We are preparing one or two articles for pullication with information on Boss' Radio Club for-

Whilst we are alway pleased to assist readera with their technical difficulties, we regret that we are unable to supply diagrams or provide instmuctions for modifying commercial or surplus equipment. We cannot supply alternative details or surplus equipment. We cannot supply aiternative details for receivens described in these pages. WER THE TELEPHONE. If a postal reply is required a stamped and addressed envelope must be enclosed with the coupon from page iii of the cover.
The Editor does not necessarily ogree with the opinions expressed by his correspondents.
mation. See also "QUA Associates", R.S.G.B. Bulletin.

学
We shall be delighted to hear from anybiody keen on this social work, and any offers of help, advice, views will be welcome from lads wherevern they are.-Ken Smith B.Sc., G3JIX ( 82 Granville: Road, Walthamstow, London, E.17).

## QSL CARDS better than wallpaper

SIR.-Correspondent C. P. Finn, of Yorkshire (P.W. January 1965 issue), asks about Electronic patterned wallpaper for his shack/bedroom.

When I had my shack/bedroom in 1932, we didn't worry about wallpaper. The main trouble was insufficient number of walls for QSL cards," blueprints, pencilled QSO's etc.
There arc plenty of abstract patterns which could soon be converted to the "Electronic" theme, but. for my choice, the old-type shack is unbeatable.P. O. Hubbart (West Croydon, Surrey).

## COMPETENT CONSTRUCTORS

SIR,-I wonder how many readers of this magazine pride themselves on being competent constructors of electronic equipment? Probably the majority, but there is, nevertheless, an increasing : number of people who, having assembled some ; transistor set, half of which has already been made in the factory and which has been prefabricated:; pre-cut, pre-aligned etc., insist on calling them: selves competent radio constructors.

Furthermore, many of these people are not even interested in how their apparatus works, but merely in the noise that it produces, and an increasing number of manufacturers are catering for the per* son with this type of outlook.

Naturally, this seems rather a snobbish attitude to take, but it would be nice to know, for the. benefit of our status as radio constructors, that we are not confused with the "step-by-step" man who only buys an f.m. tuner kit because it is cheaper than commercially available ones of the same ${ }_{p}$ quality.

I leave readers of this letter with a quote from George Orwell. "The boards I buy are ready" planed, and the legs are ready turned by the lathe. I can even go to the woodshop and buy all the parts of the table ready-made and only needing to be fitted together: my work being reduced to driving in four pegs and using a piece of sandpaper . Making a tahle will he easier and duller than peeling a potato ".-C. J. Combisiter (Cheltenham), (iloucestershire).

## LATE PASS

SIR,-As a reader of your journal since the days of Amateur Wireless (I am now 66) I fully endorse every word you say about " pirates". As you say, the RAE never stopped the real enthusiast. I know an amateur who has passed the RAE, passed the morse test, and has been on the air for a year. He is in his 70's! If this doesn't shame some of those who want the " freedom of the air", it ought to. Thank goodness these are only a minority,
I am all for encouraging the youngsters. And what a goldmine of knowledge these young people of today have in their local library-enough for them to take the MIRE examination, let alone the RAE!

Enclosed is a photograph of my den, which might be of interest. - T. W. Middleton exG2FTM (Gravesend, Kent).


Mr. Middleton's den, where equipment bosed on P.W. designs is still providing him with a hobby in his enforced retirement.

## P.W. SIGNAL INJECTOR

SIR.-With reference to the signal injector described in the August issue of Practical Wireless. after having difficulty in obtaining a 35 mm film cassette tin, I found that an "Anadin" aspirin tin was ideal for the job,-J. H. Borrowdale (Hounslow, Middlesex).

## CORRESPONDENTS WANTED

SIR,-I would like to correspond with other radio enthusiasts of my own age (13).-Edward Tweedly (27 Orchard View Drive, Kirkfieldbank, Lanark, Scotland).
SIR,-I would like to correspond with someone of my own age (14) who is interested in radio. -G. Jackson (50 Bury Street, off Sandy Lane, Stockport. Cheshire).
SIR.-I would like to hear from readers in any country who have an interest in radio and tele-vision.-S. A. Ariyasena ("Sirisewana " Nawagamuwa, W.P., Ranala, Ceylon).
SIR,-I am a regular reader of P.W. and interested in radio technology. I would like to correspond with any radio technician from Great Britain. I will answer all letters received promptly. -T. Henry C. Peiris (Watinapaha, Dewlapola, Ceylon)

RE@UESTS FOR INFORMATION ARE INSERTED IN THIS COLUMN ON THE UNDERSTANDING THAT READERS USING THE SERVICE UNDERTAKE TO REPLYTO ALL OFFERS RECEIVED AND TO RETURN ALL DATA NOT REQUIRED, BECAUSE OFTHELARGE NUMBER OF REQUESTS RECEIVED, ILLEGIBLE WRIT. ING WILL AUTOMATICALLY DISQUALIFY LETTERS FROM PUBLICATION. FOR THE SAME REASON, WE CANNOT GIVE SPACE FOR REQUESTS FOR PAST ISSUES OF "PRACTICAL WIRELESS."

## Sir - I would be grateful If any reader could sell or loan me

. circuit diagram or any information on ex-army receiver R1466, Ref. No. 10D/1409.-Brlan Wood, 99 Kinnaird Road, Sheffield 5.
.information and circuit values of ex-government photocell 0/33281 CV243 and how to make it work a relay when the light beam is cut off.-M. Gardiner, is Windmill Fields. Four Marks. Nr. Alton, Hampshire.
. the circuit or service sheet or any details of the Uitra Ultrascope oscilloscope Mk. 1.-D. CHADwtCK, 40 Abbots Road. A bbots Langley, Watford, Hertfordshire.
. any information, circuit diagram, etc., concerning the R 1392 v.h.f. receiver.-H. A. Forrester, 58 Bede Avenue, Sherburn Road Estate, Durham.
any details whatsoever on the American receiver type BC624-.....E. R. LisLe, Red Hill School, East Sution, Nr. Maidstone. Kent.
. service sheet for the Veritone "Venus" tape recorder. D. Potterill, 30 Dryden Crescent. Stevenage. Hertfordshire. wireless set 38 A.F.V.-W. B. NeEL, Landguard Lodge. Manor Terrace, Felixstowe, Suffolk.
circuit and service details of the Deleo Remy Hyat car radio, model 342.-J. Stoneman, 46 South Hall Drive, Rainham, Essex. 1154 M, - A. A. MAWHINNEY, II Donaghadee Road, Millisle, Co. Down, N. Ireland.
a circuit diagram and operating manual for a No. 38 Mk. 2 set. and a manual for a 1132 A Receiver.--S. G. Barnes. 13 Surrey Grove, Suiton. Surrey.
information concerning the modifications of the R109 Receiver (year 1943).-S. Alderton, 2 a Goldings Road, Loughton, Essex.

- information on fixing a tunable oscillator in a Bendix Receiver, BC 624A type R5019.-R. C. HART, "Brookmead" Rudwick Close, Felpham, Nì. Bognor, Sussex,
. the operating manual for the Geloso 209 Receiver.R. Buntock, 38 Bittell Road, Barnt Green, Nr. Birmingham.
... any information on the Marconi No. 9 Transmitterreceiver, and any information on the No. 46 Transmitter-receiver.-R. J. Shawe, South Supplies, 71 High Street, Poole. Dorset.
. the circuit and/or manual of the R116A communications receiver.-A. D. Couchman, 3 Manor Grove, Sittingbourne, Kent.
., , the circuit diagram and/or any other details regarding the No. 19 Set.-T. Brigos, Lindfield School, Slinden House, Slindon, Sussex.
. circuit details of the Hallicrafter S77,-C. Clakk, 94 Pensby Road, Heswall, Wirral, Cheshire.
the circuit for the PCR2 receiver.mW. Burke, GM3TQHंH, 6 Belgrave Terrace, Glasgow W2.
... the circuit diagram or details of the PCR2 receiver.G. Darling, 80 Seaton Gardens, Ruislip Manor, Middiesex. (10.. the handbook on the communication receiver DST100--R. A. Ellis, 32 e Abbey Road, St. John's Wood London N.W. 8 .
information regarding the conversion of communication receiver $1392 / 62 \mathrm{H}$ from v.h.f.la.m. to v.h.f./f.m.-J. A. Stmmson, 236 Fullwell Avenue, Clayhall Ilford, Essex.
. any information. service manual and circuits for the ex U.S. army transmitter/receiver BC620F and p.s.u.-D. W. Vitou, 169 Worthing Road, Laindon, Basildon. Essex.
the valve line-up for the Perth radio.-C. L. Dibble, c/o The Exchange Telegraph Co. Lid., 17.21 Curtain Road, London E.C. 2.
:. any information regarding the "Hersteller" b.o. type R2, battery/mains portable radio, also the circuit, etc., for the Pye 47X a.c./d.c. receiver.-R. H. Jackson, Braithwaite, Summerbridge, Harrogate, Yorkshire.
- the circuit diagram and any other information regarding the BC-624-A v.h.f. receiver,-J. S. MORTLOCK, 36 Valleyfield Road, Streatham, London S.W. 16.
the circuit for a P104 v.h.f. receiver. I cannot get this set to oscillate, and any advice on this would be most welcome. G. Spicer, 186 St. Wilfiam's Way, Rochester, Kent


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