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There are several active pastimes that depend entirely upon electronics though the participants are not necessarily involved in or even concerned with the techniques employed, but only with the resultant effects produced by some action on their part.

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So it is no wonder that radio control has a very large following and is backed by a sizeable industry catering for the special needs of these R/C enthusiasts-from models of all kinds through servo-mechanisms to complete transmitters and receivers.

Of the large numbers who participate in R/C perhaps the majority buy everything ready made and concentrate on the "real business" of operating their favourite kind of model. A fair number do however add further to their enjoyment by building their own model aircraft, boats or wheeled vehicles. And finally some, certainly a smaller proportion of the whole, actually build their own radio transmitting and receiving equipment.

To this latter group, as well as to the general electronics enthusiast, we shall be directing special attention
over the next few months. The $E E$ Radio Control System is an "entire" system and it uses a well proven circuit that is equally amenable to the needs and requirements of novice as well as experienced operator. The overall project is a result of teamwork: three designers have cooperated to produce this system which has been subjected to exhaustive field tests, culminating in the very creditable achievements by one of the trio during the Manx Soaring Championships on the Isle of Man last August.
We hope that through this project many of our readers will discover another fascinating pastime and have the additional pleasure of modestly remarking to admiring onlookers"Oh yes, I built the electronics myself".

And now for something quite different. Circuits simple, useful and all built on a standard size board. That sums up Uniboards, a new series of quick one-off's featuring commonplace discrete semiconductors that starts this month. Just the job for newcomers to cut their teeth on and assuredly worth more than a passing glance from older hands.


Our December issue will be published on Friday, November 16. See page 740 for details.


## Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications, designs or information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.
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Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.
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| T must first be emphasised that this project requires the use of an oscilloscope for the adjustments necessary to obtain the correct mark-to-space ratio for each waveform and also the shape and purity of the sinewave. This cannot be done audibly, or with an audio signal reading meter.

An audio range signal generator of this nature is a valuable piece of test equipment and has dozens of applications in testing and performance measurement of both audio and other electronic equipment. It has a total frequency coverage of 15 Hz to 100 kHz in four ranges, see Table 1.

Table 1. Band coverage of the 3-Function Generator.

| Band No. | Coverage |
| :--- | :--- |
| $1(\times 1)$ | 15 to 250 Hz |
| $2(\times 10)$ | 150 to $2,500 \mathrm{~Hz}$ |
| $3(\times 100)$ | 1,500 to $25,000 \mathrm{~Hz}$ |
| $4(\times \mathrm{kHz})$ | 10 to 100 kHz |

The output is continuously variable with maximum signal levels as shown in Table 2.

Table 2. Maximum output levels for the three functions.

| Function | Level (volts) |
| :--- | :--- |
| Sinewave | 1 (r.m.s.) |
| Square-wave | 2.5 (pk-pk) |
| Triangular-wave | 1 (r.m.s.) |

The "r.m.s." levels are according to a normal r.m.s. type a.c. (audio signal) reading meter. The maximum level square-wave signal will also read out on a similar meter at about 1.5 V (approximate r.m.s. value).


The sinewave signal has a minimum harmonic distortion factor of about 2 per cent when correctly adjusted but lower than this is not possible as the sinewave is obtained by electronic shaping within IC1 and not by pure generation.

Although the sinewave is not suitable for harmonic distortion analysis with a t.h.d. meter it is quite adequate for all audio equipment frequency response measurements, audio amplifier power output and bandwidth measurement, frequency comparison, and so on.

The triangle-wave is quite pure and also has numerous applications in electronics as well as audio, particularly as the "ramp" rise and fall is perfectly linear:

The square-wave has a rise time of only 2 microseconds and so can be used effectively for audio amplifier square-wave tests as well as for a "clock pulse" source with
a 1-to-1 mark-space ratio at any frequency within the range of the generator.

## THE CIRCUIT

The circuit diagram of the 3 Function Generator is shown in Fig. 1. Most of the work is carried out internally by the 8038 sine-square-triangle generator i.c. with external CR network-switching to provide the wide frequency coverage specified. The output signals from ICl are coupled to an opamp, IC2, with switched negative feedback to provide (a) a nominal output level of 1 V r.m.s. from the sine and triangle waves with the least possible distortion of the waveforms (b) amplified squarewave, with limited negative feedback, to obtain a fast rise time and uniform flat top characteristic, even down to 15 Hz and (c) a low


Fig. 1. The complete circult diagram of the 3-Function Generator Including the malns derived power supply. Note that R19 needs to be calculated for transformer used-see text.
to obtain a uniform output from the sine and triangle waves and a fast rise time from the squarewaves.

The output feed capacitor is kept large to obtain a flat topped square-wave down to 15 Hz but in order to check this, an oscilloscope with a d.c. input on the Y-amplifier must be used. Scope amplifier input capacitors (a.c. coupling) are usually too small in value to obtain flat top square-wave displays at frequencies as low as 15 Hz (see oscillograms in this article).

## POWER SUPPLY

The circuit requires a smooth 30 V d.c. supply. This is derived from the mains in a conventional manner. Mains voltage enters the unit via S3 and appears across the primary of T1; 24 V a.c. (nominal) is produced across Tl secondary which is then full-wave rectified by the diode bridge D2 to D5, producing a d.c. level across Cl 0 equal to the peak value of Tl secondary voltage (i.e. $24 \sqrt{ } 2$ ) plus an overvoltage due to the regulation factor of the transformer.
The prototype used a transformer with a secondary current rating of 250 mA , resulting in total voltage at $\mathrm{ClO}+\mathrm{ve}$ of 41 V . The current required by the circuit is 25 mA . Therefore to obtain 30 V at $\mathrm{Cll}+\mathrm{ve}, 11$ volts must be dropped across R19 when 25 mA flows.

From Ohm's law, R19 = 11/0.025 $=440$ ohms. The nearest preferred value above is chosen, i.e. 470 ohms.

To determine the value of R19 for other transformers that might be used, carry out the following.

With the power supply section not connected to the rest of the circuit, measure the voltage across ClO , $\left(V_{\mathrm{m}}\right)$ and then calculate the value of a resistor, $R_{p}$, to place in parallel with C10 to cause 25 mA to flow:

$$
R_{\mathrm{p}}=V_{\mathrm{m}} / 0.025 \mathrm{ohms}
$$

Measure the voltage now at Clo +ve , call this $V^{\prime}$ m. Remove $R_{\mathrm{p}}$. The value of $\mathrm{R} 19=\left(V_{m}^{\prime}-30\right) / 0 \cdot 025$ ohms.

Calculate R19 wattage from $\left(V_{m}^{\prime}-30\right) \times 0.025$ watts.


The complete generator and its power supply will fit comfortably into a Verobox type $75-1412 \mathrm{~K}$ which has aluminium front and rear panels. The generator circuit board and its controls are situated on the front panel with the rear panel holding the power supply board and transformer.

Drilling details for the front panel are shown in Fig. 2. The diameter of some of the holes, e.g. the on/off switch and the panel

Resistors

## 

| R1 | $1 \mathrm{k} \Omega$ | R7 | $15 \mathrm{k} \Omega$ |
| :--- | :--- | :--- | :--- |
| R2 | $22 \mathrm{k} \Omega$ | R8 | $150 \mathrm{k} \Omega$ |
| R3 | $4 \cdot 7 \mathrm{k} \Omega$ | $R 9$ | $33 \mathrm{k} \Omega$ |
| R4 | $10 \mathrm{M} \Omega$ | R10 | $3 \cdot 9 \mathrm{k} \Omega$ |
| R5 | $4 \cdot 7 \mathrm{k} \Omega$ | R11 | $1 \mathrm{M} \Omega$ |
| R6 | $15 \mathrm{k} \Omega$ | R12 | $1 \mathrm{M} \Omega$ |

All $\frac{1}{4}$ watt carbon $\pm 5 \%$ except where
stated otherwise

| R13 | $10 \mathrm{k} \Omega$ |
| :--- | :--- |
| R14 | $10 \mathrm{k} \Omega$ |
| R15 | $10 \mathrm{k} \Omega$ |
| R16 | $15 \mathrm{k} \Omega$ |
| R17 | $120 \mathrm{k} \Omega$ |
| R18 | $680 \Omega$ |
| R19 | $470 \Omega$ |
|  |  |
|  | text) |

Potentiometers
VR1 $10 \mathrm{k} \Omega$ carbon log.
VR2 $1 \mathrm{k} \Omega$ miniature horizontal preset
VR3, $4 \quad 100 \mathrm{k} \Omega$ miniature horizontal preset (2 off)
VR5 10k carbon linear.

## Capacitors

| C1 | $0 \cdot 1 \mu \mathrm{~F}$ ceramic or plastic | C7 | $500 \mu \mathrm{~F} 25 \mathrm{~V}$ elect. |
| :---: | :---: | :---: | :---: |
| C2 | $0.22 \mu \mathrm{~F}$ ceramic or plastic | C8 | $10 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. |
| C3 | $0.022 \mu \mathrm{~F}$ ceramic or plastic | C9 | $100 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. |
| C4 | 2200pF ceramic or plastic | C10 | $500 \mu \mathrm{~F} 50 \mathrm{~V}$ elect. |
| C5 | 500 pF silvered mica | C11 | 2,200 $\mu$ F 35V elect. |
| C6 | $10 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. | C12 | $0 \cdot 0047 \mu \mathrm{~F}$ ceramic or plastic |

Semiconductors

| IC1 8038 | function generator i.c. |  |
| :--- | :--- | :--- |
| IC2 | 741 | operational amplifier (8 pin d.i.l.) |
| D1 | 1 N 4148 | or similar small signal silicon diode |
| D2 to | D5 | 50 V |
| 1 A bridge rectifier |  |  |

Miscellaneous


S1 1-pole 4-way rotary switch
S2 2-pole 3-way rotary switch
S3 mains single-pole on/off toggle
T1 mains primary/ 24 V 25 mA secondary-see text
LP1 mains panel mounting neon
SK1 panel mounting phono socket
0.1 inch matrix perforated board size $58 \times 26$ holes; 3-way terminal block; three-core mains cable; 8 -pin and 14 pin d.i.l. sockets ( 1 off each); control knobs (4 off); case Verobox type $75-1412 \mathrm{~K}$; plastic spacers and 6BA fixings; rubber grommet to suit mains cable; 6BA solder tag; connecting wire; 1.5 mm thick clear Perspex approx. $100 \times 75 \mathrm{~mm}$.


Fig. 2. Dimensions and drilling details for the front panel of the unit with some suggested panel markings.
mounted neon may need to be changed to suit the components used.

The generator circuit is built on a piece of 0.1 inch matrix perforated board size $58 \times 26$ holes. The layout of the components and wiring details on both sides of the board and interwiring between the panel mounted controls is shown in Fig. 3. Although the generator circuit board layout is not critical, the constructor is advised to retain the position of all the components as closely as possible to avoid interaction and waveform crosstalk.

In the prototype the generator circuit board was mounted on the front panel using 30 mm long plastic spacers and self-tapping screws.


Fig. 3. Above. The layout of the components and interwiring on both sides of the circuit board. Note that some wires pass through holes to make connections to components/wiring at other locations. Left. The internal face of the front panel showing component positions, interwiring and connections to be made to the board.


The completed generator circuit board showing positioning of components.

There is no particular order to be followed in the assembly of components on this board except perhaps to begin by inserting the Veropins, used for component anchorage, and the i.c. sockets. Some interconnecting wires use tinned copper wire and others use p.v.c. covered wire. Where there is any danger of a link wire touching another wire or component lead, the p.v.c. type is essential. Alternatively, tinned copper wire and sleeving may be employed.

When assembly is complete, sufficient lengths of flying lead should be attached to the board to reach the panel mounted components, and this wiring carried out. The board can then be screwed in place on its spacers.

## POWER SUPPLY SECTION

As previously stated, the power supply board is fitted with the transformer on the back panel. The board is mounted on spacers to keep it clear of the metal back panel. The layout of the components is shown in Fig. 4 together with the interwiring on the underside. Note that this differs slightly from that in the photograph. The author used two $1,000 \mu \mathrm{~F} 25 \mathrm{~V}$ capacitors in series to form a $500 \mu \mathrm{~F}$ 50 V capacitor. This has been replaced in the text and diagrams by a single capacitor of this value.

Secure the transformer to the back panel remembering to place a solder tag on one of the fixings for earthing purposes. Make the connection between the board and Tl secondary and secure the board in place; R19 should not yet be connected, its value may need to be determined as explained earlier.

The two wires interconnecting the boards should not be connected until R19 is in place. In the prototype, a convenient method of connecting the mains cable to the transformer primary was to use a short length of plastic screwterminal. The mains cable should of course be passed out through the rear panel via a rubber grommet or strain relief bush. Complete the wiring to S3 and LP1.

## CALIBRATED FREQUENCY SCALE

A full-size copy of the calibrated frequency scale as used on the prototype is shown in Fig. 5. This can be cut out or photocopied and pasted on thin card. It is secured under the locknut of the frequency control VRl but if a thin Perspex plate can also be made to cover it, so much the better.

A graticule type pointer, like that on the prototype is also worth while and not difficult to make from thin Perspex to the size
shown in Fig. 6 and which is glued (Araldite) or screwed to the back of a plain control knob

## CHECKING OUT AND ADJUSTMENT

As already mentioned, an oscilloscope is necessary for adjustment of the presets VR2, VR3 and VR4. The oscilloscope Y-amplifier input is connected to the generator output socket and the output control set at maximum. A preliminary check with the frequency range switch S 1 on $\times 10$ frequency scale pointer at $100(1,000 \mathrm{~Hz})$ and waveform switch S 2 set to "square", will establish that the generator is operating, in which case first check the supply rail positive to ground voltage at the junction of R19 and C11 (power supply) which should be $30 \mathrm{~V} \pm 1 \mathrm{~V}$. If not, it may be necessary to slightly change the value of R19 to obtain 30 V as close as possible otherwise the output level and calibration of the generator may be affected. With this done, adjust VR1 to obtain a square-wave (still at $1,000 \mathrm{~Hz}$ ) with a uniform l-to-1 mark-space as in the oscillogram Fig. 7b. Next, switch S2 to sinewave output and adjust VR3 and VR4 together, each a little bit one



Mounting of the generator circuit board on the rear of the front panel.



Fig. 7. Photograph of oscilloscope screen containing the three functions generated by the prototype (a) sinewave (b) squarewave (c) triangle-wave.
way or the other, to obtain the closest possible replica of the sinewave in oscillogram Fig. 7a. Each of the waveforms shown in this photo were taken from the prototype generator.

Now switch to triangle wave for which no other adjustment is necessary as its mark-space has already been established. It will appear as in the oscillogram Fig. 7c.

If an r.m.s. reading a.c. voltmeter is available check that the output level is appropriate from each waveform and over the whole frequency range. A reasonable assessment of this can of course be made with a calibrated oscilloscope.

If a distortion analyser is available, the sinewave purity can be adjusted with VR3 and VR4 until the lowest possible distortion i.e.,


Fig. 8. Shows the rise time obtained from the prototype square-wave output signal.
about 2 per cent at $1,000 \mathrm{~Hz}$ is obtained.

Some further checks on squarewave outputs can be carried out with a calibrated oscilloscope and preferably one with a d.c. input to the Y-amplifier and a time base range in the microseconds region. On a fast time base range the rise time of the square-wave can be verified and this should be in the region of 2 microseconds for 90 per cent of the rise as shown in the oscillogram Fig. 8.

At 15 to 20 Hz the square-wave should have an almost perfectly flat top as in oscillogram Fig. 9 but this will only be apparent with d.c. coupling into the 'scope.

## USES

An audio range three waveform generator of this nature is a very desirable item of test equipment


Fig. 9. Even at 15 Hz the top of the squarewave is almost flat.
but its full use requires an a.c. (audio range) voltmeter and an oscilloscope at least to carry out tests and measurements on audio amplifiers, tape recorders and various kinds of purely electronic circuitry as mentioned earlier.

With the extra essential items of test equipment as above, one could measure frequency responses of audio amplifiers and tape recorders, responses of tonecontrols, filters and pre-amplifiers, carry out square-wave tests on amplifiers, check frequencies of other generators and oscillators and measure the power output of audio amplifiers etc.

Incidentally a quite good but secondhand oscilloscope is not difficult to get hold of at a reasonable price and is one of the most valuable of all the numerous items of test equipment to be found in workshops and laboratories. I


## EE CROSSWORD No 21 avo, en nwtom

## ACROSS

1 Wound-up instrument found in magnetic fields $(6,4)$.
5 Rough and offensive.
7 We would usually not be clapped out of these conductors.
8 Analysis of integrated circuit system goes through the body.
10 Part of the less-heavy e.m spectrum (5, 4).
13 Something fishy about this transistor output.
14 Wires with electromagnetic privacy $(8,5)$.
16 To r.m.s. about will give a rough passage (Anag.)
17 Chronological list for TV and calculator.
19 Single-minded tape (3,5).
21 Carrier on the waves.
22 Horsey problem.
23 Maximum displacement across a wave $(4,2,4)$.

## DOWN

1 Defensive device against electrical disturbance.
2 No lag beside (Anag.).
3 It's no sin.
4 It's a sort of output as a matter of debate.
5 Random access memory, sheepishly.
6 To break up an electronic marriage?
9 A test or I'm guessing (Anag.).
$11 C R$ might define one $(4,5)$.
12 Sound intensity measurer (1, 1, 5).
14 End of transmissions for the day $(4,4)$.
15 The table is turned into decorative activity.
17 Lagging behind or leading, we all pass through one from time to time.
18 --|---|.-•|.../.


20 Intellectual head-characteristic.
21 Beat frequency oscillator, to begin with.
 BY F.G.RAYER

ATUNER to provide a.m. reception on medium and long wave bands increases the scope and entertainment value of an audio amplifier. This tuner has sufficient output for even insensitive amplifiers, while avoiding the relative complication of a superhet circuit. Coverage is approximately 1600 to 600 kHz m.w., and 490 to 185 kHz l.w., or 360 to 145 kHz l.w.

## CIRCUIT DESCRIPTION

The circuit diagram of the tuner is shown in Fig. 1.

The circuit comprises an r.f. amplifier, TR1, a diode detector, D1 and an audio amplifier TR2 with high output impedance.

Signals generated in the aerial
are fed via SK1 into Ll primary and induced into Ll secondary to reach the gate of the r.f. amplifier, TR1. The potentiometer VR1 is the gain and volume control. As the wiper of VRl is moved towards Ll pin 5, the source bias is increased thereby reducing the gain of TR1. The aerial signal in Ll primary is attenuated at the same time.

The drain terminal of TR1 is coupled to the primary of coil L2, pins 5 and 6 , which is tuned by the second section of the ganged capacitor, Clb. Each section has its own trimmer, C2 and C5 respectively.

Note that a dual 500pF gang can be used but will require a little extra space.

A tapping on the secondary winding of L2 is used as a signal source for the detector/smoothing capacitor D1/C6. A tapping is used to avoid unnecessary loading of L2.

The audio output from D1 is coupled to the base of audio amplifier TR1 by C7; TR2 is wired as a



Fig. 1. The complete circuit diagram for the Medium and Long Wave Radio Tuner.




Fig. 3. Construction of a simple tuning pointer disc. Half of a brass spindle coupler is glued to a 70 mm diameter Perspex disc with an engraved radial line.
common emitter amplifier providing considerable boost. The output is at the collector which is capacitively coupled to the output socket SK3.

## TUNING COILS

The tuner uses two identical coils, having six tags, see Fig. 2. Count from the tag ring slot. Bandswitches Sla and Slib are sections of a slide switch, and short out both of the longwave windings (pins 3 and 4) for medium wave reception.

The coils are of fixed inductance, and do not have adjustable cores. The only adjustment necessary will be to the trimmers C2 and C5. To do this, tune in a signal near the high frequency end of the m.w. band (ganged capacitor nearly fully open) and rotate C2 and C5 for best results.


## COMPONENT BOARD

Most of the components including the dual-ganged capacitor are assembled on a piece of 0.15 inch matrix perforated board, $30 \times 12$ holes, as in Fig. 2. The 2-gang capacitor used has two threaded holes, so that the board can be fitted to it with short 4BA bolts.

First solder a lead to the capacitor rotor or frame tag, and bring it down through a hole in the board. This is the earthing point MC in Fig. 2. Extra washers or similar means of spacing about 3 mm thick will be needed between this capacitor and the board.

Assemble and interconnect the board-mounted components according to Fig. 2. In many places the wire ends of components can reach to the required points. Elsewhere, 22 s.w.g. or similar connecting wire is recommended.

Prepare the front panel to accept the panel-mounted controls and secure these and the board in position.

The ganged capacitor is fitted to the panel by means of three 12 mm long 4BA bolts, with two nuts each,

to lock against the capacitor and panel. This capacitor provides sufficient support for the board.

Should a capacitor without slow motion be fitted, then this can come nearer the panel as the spindle will be shorter. Take care that the bolts are not so long that
they project inside the capacitor.
The aerial and earth sockets, SK1, SK2 are to be mounted on a small piece of Paxolin or similar material to be fitted to the rear of the case.

Complete the interwiring as shown in Fig. 2.

Resistors

| R1 | $100 \mathrm{k} \Omega$ |
| :--- | :--- |
| R2 | $100 \mathrm{k} \Omega$ |
| R3 | $220 \Omega$ |
| R4 | $22 \mathrm{k} \Omega$ |
| R5 | $1 \cdot 8 \mathrm{M} \Omega$ |
| R6 | $4 \cdot 7 \mathrm{k} \Omega$ |
| All | $4 W$ carbon $\pm 5 \%$ |

Capacitors
C1 $2 \times 365$ pF dual ganged (Jackson type 0), slow motion (preferred)
C2 60 pF compression trimmer
C3 $0.01 \mu \mathrm{~F}$ ceramic or plastic
C4 $0.01 \mu \mathrm{~F}$ ceramic or plastic
C5 60 pF compression trimmer
C6 220pF ceramic or plastic
C7 $0.47 \mu$ F ceramic or plastic
C8 $4 \cdot 7 \mu \mathrm{~F} 6 \mathrm{~V}$ elect.
C9 $100 \mu \mathrm{~F} 12 \mathrm{~V}$ elect.


Semiconductors
TR1 CA40673 or 3N201 dual gate silicon n-channel MOSFET TR2 BC108 silicon npn
D1 OA81 or similar germanium diode
Miscellaneous
L1, L2 Repanco type DRR2 (2 off)
VR1/S2 5 kilohm carbon linear with ganged switch
S1 d.p.d.t. slide switch
SK1, 24 mm in sulated sockets or similar (2 off)
SK3 3.5 mm jack socket
B1 9V PP3 or any other 9 V battery
Circuit board; 0.15 inch matrix perforated board, size $30 \times 12$ holes; battery connector to suit B1; knobs (2 off); 4BA and 6BA fixings; Perspex and bush (for dial); Paxolin or similar, $100 \times 40 \mathrm{~mm}$ (to hold SK1, SK2); case $150 \times 100 \times 100 \mathrm{~mm}$.

Trimmers C2 and C5 are soldered directly to Cla and C 2 b , and their second tags supported by a short, stout wire to the gang frame. They are almost vertical, so that they can be adjusted by means of a small screwdriver from behind, with the tuner in its case.

The case employed in the prototype had dimensions $150 \times 100 \times$ 100 mm internally, and was made of metal; plastic or thin wood could also be used.

## POWER SUPPLY

Current drain is small (3mA measured) and an internal 9 volt battery is therefore suitable.

The tuner may be operated from a well decoupled and smoothed supply obtained from the main audio amplifier, of about 9 to 12 volts, with negative earth.

## TRIMMING

Initially set the trimmers to near maximum capacity. Subsequently adjust both for best reception of
a medium wave transmission near the h.f. band end (say 1600 to 1400 kHz ) as mentioned. For optimum results adjust C2 with the actual aerial and earth which will be used, already connected.
For the alternative l.w. band mentioned, cores may be obtained which can be screwed to the l.w. winding ends of the coils. These are not necessary, however, for 200 kHz reception.
The aerial can be a few feet of wire indoors, or a somewhat longer indoor wire carried along one (or possibly two) walls of the room, near the ceiling. Either a short or rather longer out-door aerial may be used if available. It can be worthwhile to try one or two alternatives.
It is recommended that an earth connection be provided if feasible.

## IN USE

The usual type of screened lead should be employed to feed the audio signals from the tuner to the main amplifier.

If the tuner is used for personal headphone listening, a high resistance headset approximately 2 kilohms will be most satisfactory.

Note that the values are so arranged that the maximum possible gain setting of VR1 brings the tuner to the point of regeneration, as this allows improved sensitivity. This was found to cause no difficulty with an earth provided, but with no earthing VRl must be adjusted accordingly, or resistor R3 increased in value until it is just impossible to bring TR1 into oscillation. A resistor of about 1.5 kilohms should be suitable.

A tuning pointer can be made from a stout wire soldered to the capacitor spindle, or, as used in the prototype, a disc of thin Perspex about 70 mm in diameter can be fitted to a bush with set screw, obtained from an old control knob with a line scribed along a radius of the disc. A 180 degree scale can be glued to the front panel behind the Perspex disc and later calibrated.

## 

## I.C. SOCKET

For sometime I have been using one of those $T$-Dec breadboards and to use a d.i.i. integrated circuit with this you need a special adaptor. This can cost between $£ 2$ to $£ 2 \cdot 50$.
I first came up with the idea of an i.c. socket fitted with flying leads, but this proved to be a bit clumsy. Then I hit upon an idea of using a wirewrap i.c. socket, bending the pins as shown in the diagram and then snipping the ends level. The socket can then be plugged in and out of the T-Dec with ease.


The cost of the Wirewrap I.C. socket should be between 20 and 30 pence, which is a considerable saving on the original.
L. A. Privett, Barking.

## The Adventures of Tanty Bead




Semiconductors is a term that embraces a very important family of electronic devices. The most widely used, and best known, of such devices are the diode and the transistor.

The simplest semiconductor device is the diode. This functions as a oneway device (or "valve") for electronic current. It has two terminals or leadout wires. One connection is distinguished by a mark of some kind on the body of the component, and this is the cathode. For normal conduction this must be connected to the negative side of the circuit in which it is to be used. The other (usually unmarked) is the anode and goes to the positlve side of the circuit. See Fig. 1.

There are a variety of diodes, varying size, shape and form. See Fig. 2.

One kind of device commonly encountered in electronic circults is that generally known as a general purpose signal diode. Many of these resemble a small resistor in outward appearance and have a coloured band at one end of the body which identifles the cathode.

## A SPOT OF CONFUSION

Other types of diodes have some other kind of mark adjacent to the cathode lead.

Perhaps somewhat confusing is the use of a + sign or a red band or tip to denote the cathode on certain diodes. This is a throw-back to earlier days when diodes were used chlefly for power rectification. The positive

## FOR




Fig. 1. Circult symbol for a diode, with cathode (k) and anode (a) identified.


Fig. 2. Typical digdes and methods of Indentifying the cathode.
side of the direot current (d.c.) output from a power reotifying circuit comes from the cathode of the diode rectifier, and so this method of coding makes sense. But when the diode is employed in other circuit arrangements the basic logic of this method of identification is somewhat obscure and confusion is frequently caused.

## CIRCUIT SYMBOL

The symbol normally used for a diode in circuit diagrams is shown in

Fig. 3. (a) Diode forward-blased-conducts. Can be consldered as a switch that is closed. (b) Diode reverse-blased-does not conduct. Can be considered as a switch that is open.


Fig. 1. The "bar" represents the cathode. The arrow head represents the anode and points in the direation of conventional curent flow, that is positive to negative.

It has been general practice to mark the cathode of the diode symbol with an " + ". But because of the possible confusion previously referred to, we have abandoned this and now mark the two ends of the diode symbol " $k$ " and " $a$ ", representing cathode and anode respectively.

## DIODE OPERATION

The diode conducts when the anode (a) is connected to the positive point of a circuit, and the cathode ( $k$ ) to a more negative (less pasitive) point. See Fig. 3 a.

When connected the other way round, (or if the circuit voltages reverse) the diode will not conduct, but becomes a complete barrier to the flow of direct current. See Fig. 3b.

When the diode is used to rectify alternating current (a.c.) it behaves like a switch, "opening" and then "closing" as the a.c. changes direction, that is swinging from positive to negative. See Fig. 4.

The unidirectional output from the diode is a series of positive going pulses. The negative-going half of the a.c. input is suppressed.

Fig. 4. Diode used as a rectifler of a.c. $T$ is a malns transformer. When " $x$ " is positive, the diode will conduct, and d.c. (conventional current) flows from cathode back to other end of transformer winding. When " $x$ " is negative, the diode will not conduct.



The start of a new series of six easy-to-build transistor-based projects. All use a standard size piece of stripboard, 10 strips by 24 holes.

THIS simple single-transistor circuit is designed to sound a miniature audible warning device when light falls on to a photocell. The photocell is normally mounted in a dark room and the alarm is triggered when either the room lights are switched on or possibly when light from an intruder's torch falls directly on to the photocell.
The circuit will operate satisfactorily from a 9 volt battery but as it is probable that the device will come in for regular use the device described here can be wired to operate from the "9 Volt Power Pack" project to be described later in this series.

## CIRCUIT DESCRIPTION

The circuit diagram of the Opto Alarm appears in Fig. 1. The photocell, PCC1 is an ORP12 light-dependent resistor which is located in the room to be protected, and is connected by means of PL1 and SK1. Together with R1, PCCl forms a potential divider: the voltage at the junction of R1 and PCC1 varies with the amount of light striking the l.d.r.
In absolute darkness the resistance of an ORP12 is at least 10 megohms, and so the voltage at the junction of $\mathrm{Rl} / \mathrm{PCCl}$ is very nearly that of the supply rail, 9 V . Transistor TR1 is therefore firmly switched off as its base is not biased.

When light falls on PCC1, its resis. tance drops (albeit relatively slowly) and this causes TRI to switch on. A
triggering pulse is therefore delivered to the gate of CSR1 and this component conducts. The audible warning device (WD1) will therefore sound.
The thyristor will now remain in this low impedance state even if the triggering signal is removed. The only way to reset CSR1 and mute the alarm is in this case to switch off the mains power supply, or switch off the battery if dry cells are used instead. Resistor R5 will ensure that a minimum holding current is flowing in the anode-cathode circuit of the triggered thyristor, and so preventing any undesirable resetting.

## BUZZER

It is important to note that conventional electromechanical buzzers should not be used in this circuit. They feature a very high current consumption normally, and apart from destroying the specified thyristor such a unit could greatly reduce battery life if the circuit is powered by conventional batteries. The miniature audible warning device used here has a current consumption of only 15. 20 mA .

Whilst the response time of the l.d.r. is relatively slow, experimentation with resistor values enabled a design to be produced which reacts quickly to a change in light: the alarm is triggered, for example, by a torch beam skimming over the photo-resistor in a darkened room.

Finally, C1 and C2 decouple the power supply and prevent triggering of the thyristor during initial switch. on, A 9 volt supply is connected via SK2, the tip of the jack plug being +9 V as utual.


The prototype was built into an ABS "Bimbox" type 4003. This measures approximately $85 \times 55 \times$ 35 mm and has a steel front panel. The circuit can be accommodated neatly on a piece of 0.1 inch matrix stripboard, 10 strips by 24 holes.

There should be no problems with the construction of the circuit ${ }^{\text {Fig. }} 2$ illustrates the recommended arrangement of components. As usual note carefully the connections to the semiconductors and in particular ensure the correct polarity of Cl .

The metal panel of the box is drilled to carry the miniature buzzer and also the two jack sockets. A small hole is also required to enable the leadouts from the bleeper to pass through the metal panel to the circuit board inside.

All interconnections between the component board and front panel can be completed with stranded flexible hook-up wire. Make quite certain that both jack sockets are wired the right way round. Both sockets must be wired exactly as shown: note that the metal panel will in fact be connected to 0 V through the jack sockets.

## LIGHT SENSOR

The photocell arrangement in the prototype is shown in Fig. 2. The ORP12 is mounted upon a small piece of tagstrip and connected to its respective jack socket using twin-core flex terminated with a 3.5 mm jack plug. The length of the flex can be in excess of 5 metres.
No setting up is required, simply mount the l.d.r. in the room to be monitored. Obviously it should not be obscured by any object in the room.
One final point is to remember to connect up all jack sockets before switching on the power. If this is not done then there is the possibility that the " 9 Volt Power Supply" (if used) could be shorted out when the jack plug connecting it is being inserted into the jack socket.
If battery operation is required, the power input socket SK2 should be replaced by an on/off switch located so as to allow a PP3 battery to sit in the case.


## COMPONENTS

## Resistors

| Resistors |  |
| :--- | :--- |
| R1 | $22 \mathrm{k} \Omega$ |
| R2 | $4.7 \mathrm{k} \Omega$ |
| R3 | $680 \Omega$ |
| R4 | $2.2 \mathrm{k} \Omega$ |
| R5 | $1 \mathrm{k} \Omega$ |

## See

R2 $4 \cdot 7 \mathrm{k} \Omega$
R4 $2 \cdot 2 \mathrm{k} \Omega$
All W
All $\ddagger$ W carbon $\pm 5 \%$
page 742

## Capacitors

C1 $150 \mu \mathrm{Fl} 6 \mathrm{~V}$ elect.
C2 $\quad 0 \cdot 1 \mu \mathrm{~F}$ polyester C280 or similar
Semiconductors
TR1 BC178 silicon pnp
CSR1 MCR102 thyristor rated 30 V 0.8 A or similar

PCC1 ORP12 or similar light dependent resistor

Miscellaneous
SK1, 23.5 mm jack socket (2 off)-

## see text

PL1 3.5 mm jack plug
WD1 miniature 9 V audible warning device
Stripboard: 0.1 inch matrix, 10 strips $\times 24$ holes* ; case BIM 4003 or similar; tagstrip; twin-core flex; stranded connecting wire; 6BA fixings including 5 mm spacers; Optional components, 9 volt battery and connec tor; on/off switch.

- Avallable in packs of five boards.

Approx. cost Guidance only
E22.00 excluding case

FIg. 1. The clrcult diagram of the Opto Alarm. The dotted components replace SK2 for an integral battery version.


Fig. 2. The layout of the components on the topside of the stripboard and the breaks to be made along the copper strips on the underside and interwiring between board and panel mounted components. Left shows the l.d.r. fixed to a tag strip enabling the l.d.r. to be mounted and connected to a jack plug to sult SK1.


The ratio of the voltage to the current we call the "resistance". The mathematical way of defining resistance is by the equation $R=V / I$
where $R$ is the resistance, $V$ the voltage and $I$ the current. We call this equation Ohm's Law after its discoverer.

The units of resistance are ohms, one ohm being the resistance which allows one ampere to flow when one volt is applied. Conversely we can say that one ohm produces a voltage drop of one volt when one amp flows through it.

## CURRENT VERSUS VOLTAGE GRAPH

Another way of visualising resistance is by plotting a graph of current against voltage in a given component. The resistance is then given by the slope of the graph.

A pure resistance gives a straight line current versus voltage graph-we say there is a linear relationship between current and voltage, see Fig. 2.1.

Other components may not give a straight line but we can still find the resistance at any point on the graph by drawing a tangent to the curve and then measuring the slope of this line.

## SWITCHES

A switch can be defined as a twostate device-in one state it has extremely high resistance (it is an insulator), and in the other state it has very low resistance (it is a conductor).
The force which causes it to change state may be mechanical, as in an ordinary light switch, or an electric current or voltage, as in the case of a relay or an electromechanical solid-state switch.
Switches vary in their specifications as to how much voltage they can withstand in their insulating or "off" state, and how much current they can carry in their conducting or "on" state.
Switches can have more than just two contacts which are either open or closed. Mechanical switches with eight or more contacts are not uncommon.
A very useful type of switch is the changeover type where a moving contact, or wiper, makes contact with either one terminal or another. This type of switch can be used as a normaliy closed
switch, a normally open switch, or can be used to switch from one voltage to another.
The circuit symbols for various types of switch are shown in Fig. 2.2.

## RESISTORS

Perhaps the most common circuit element is the resistor. Resistors come in a variety of shapes and sizes but they all have a common function-to accurately set current levels in a circuit when given voltages are present.
Resistors are somewhat taken for granted in electronic circuits but it is quite remarkable that a component can give such predictable behaviour over a vast range of applied voltages.
Early resistors tended to be large rods of carbon even for quite low power dissipations. This was because internal heating was a problem in the solid type of construction. Modern resistors use sophisticated techniques to give very high performance and stability combined with small physical size.
The circuit symbols for various types of resistors are given in Fig. 2.3.

## TYPES OF RESISTOR

The actual resistive part of a resistor can be carbon, a thin film of metal or metal oxide, or a wire made of a suitable alloy. The cheapest and probably the most widely used are carbon type but often, especially in precision instruments, the shortcomings of this type of resistor necessitate the use of more expensive metal film or metal oxide resistors.
The quality of a type of resistor can be judged in two ways: its tolerance and its stability with changes in temperature, humidity, etc. The concept of tolerance is, perhaps, a new one and therefore requires some elaboration.

## TOLERANCE

When resistors (or any component for that matter) are actually produced, the manufacturer cannot ever make his components exactly match the nominal specification of that component. He must compromise between accuracy and cost so he does not attempt to make resistors of
exactly the resistance required but, instead, specifies a band of values around the nominal within which the component is acceptable. In general, the closer the limits of acceptance are to the nominal value, the higher the cost.

The band around the nominal value is usually specified in terms of a percentage. A "ten ohm, five per cent" resistor is therefore a resistor whose real value can be anything from $9 \cdot 5$ ohms to $10 \cdot 5$ ohms.

Typical tolerances for resistors are 20 per cent, 10 per cent, 5 per cent, 2 per cent and 1 per cent. Tolerances of one per cent or better make the resistor what is called a "precision" resistor.

In general, the designer likes to produce circuits where low tolerance (high percentage) resistors can be used since this keeps down costs. However, there are many instances where close tolerance (low percentage) resistors are essential.

The concept of tolerance has led to the formulation of a range of values for resistors which all manufacturers now follow. These values are called preferred values and the way the actual values have been arrived at is quite interesting.


Fig. 2.1. Plotting current against voltage shows there to be a straight line (linear) relationshlp between the two. The slope of the graph gives the resistance.


Fig. 2.2. Circuit symbols for switches. (a) shows a simple on/off type; (b) a changeover, and (c) a double-pole changeover.

## PREFERRED VALUES

Since manufacturers cannot make a resistor of every value imaginable, they have arrived at a set of values which the designer can choose from. This obviously puts constraints on the circuit which the designer must be aware of.

We said earlier that a "ten ohm, five per cent" resistor could take any value up to $10 \cdot 5$ ohms. There is thus no point in making a resistor whose nominal value is less than this. So, what is the next highest value that he should make?

The lower limit of the tolerance band of the new resistor should not overlap with the upper limit of the "ten ohm" resistor. A little calculation shows that the next value is 11 ohms (to the nearest whole number). Using the same principle we can find the next highest value which turns out to be 12 ohms.

Continuing in the same way, a whole string of values can be found up to 100 ohms. Above this the values are simply ten times the previously calculated values.

It turns out that for five per cent resistors there are 24 values between 10 and 100 ohms. We call any set of values where the upper limit is ten times the lower limit


Fig. 2.3. Circuit symbots for resistors. (a) shows a simple resistor, (b) a light dependent resistor (I.d.r.) and (c) a thermistor.


Practical examples of the components depicted In Fig. 2.3. (a) resistor (b) light dependent resistor and (c) thermistor.
a "decade", so the previous statement can be summarised by saying that there are 24 values per decade. The values are all listed in Table 2.1 along with the other series for 20 and 10 per cent.

Each of these series is called an " $E$ " series and to denote the particular one we mean, we follow the $E$ with the number of values per decade. Hence Table 2.1 lists the E6 (20 per cent), E12 ( 10 per cent) and E24 (5 per cent) series.

## POWER RATING

When we looked at conduction in solids we saw how electrons move-bouncing around in a random manner but with an overall drift against the field. The collisions which occur generate heat and the greater the current the more collisions occur.

Each collision therefore means that the electron loses some of its energy as heat. We say that power is dissipated when current flows in a resistive element.
The amount of power dissipated is proportional to the current flowing through, and the voltage across the resistor. Thus

$$
P(\text { power })=
$$

$I$ (current) $\times V$ (voltage)
Heat will be dissipated in any resistive element in a circuit whether it be an actual resistor or a piece of wire, since this is bound to have some resistance at normal temperatures.

When resistors are designed, the manufacturer tests how much power the type of resistor can dissipate without any damage. If too much current is passed through a resistor it will get hot and eventually burn out. Thus when a resistor is given a power rating it is really a summary of the maximum voltage and current which the resistor can withstand.

To calculate these two quantities from the power rating and the value of the resistor, we must return to Ohm's Law.

If a voltage $V$ is placed across a resistance $R$ then the current $I$ is given by

$$
I=V / R
$$

Now we have seen that

$$
P=V \times I
$$

so, substituting in this equation we get

$$
P=V \times V / R \text { or } P=V^{2} / R
$$

Rearranging we get

$$
V=\sqrt{ }(P \times R)
$$

TABLE 2.1
Range of Preferred Values for Resistors

| Tolerance | Series | Values per decade |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 20\% | E6 | 10 |  |  |  | 15 |  |  |  | 22 |  |  |  |
| 10\% | E12 | 10 |  | 12 |  | 15 |  | 18 |  | 22 |  | 27 |  |
| 5\% | E24 | 10 | 11 | 12 | 13 | 15 | 16 | 18 | 20 | 22 | 24 | 27 | 30 |
|  | E6 | 33 |  |  |  | 47 |  |  |  | 68 |  |  |  |
| 10\% | E12 | 33 |  | 39 |  | 47 |  | 56 |  | 68 |  | 82 |  |
| 5\% | E24 | 33 | 36 | 39 | 43 | 47 | 51 | 56 | 62 | 68 | 75 | 82 | 91 |



Typical resistors (left) a wire-wound 5 watt fixed resistor, and (below) carbori resistors ranging from 1 watt to $1 / 10$ th watt. (Top right) three variable resistors (potentiometer): standard control type and two miniature presets. (Bottom right) light dependent resistor and thermistor.

Let us look at a real case. What is the maximum voltage which we can safely apply across the 100 ohm, one watt resistor?
$V=\sqrt{ }(P \times R)=\sqrt{ }(1 \times 100)=10 V$
We can find the maximum current by substituting

$$
V=R \times l \text { in } P=V \times I
$$

giving

$$
P=R \times I \times I \text { or } P=R \times I^{2}
$$

Rearranging, $I=\sqrt{ }(P / R)$
Again, let us look at a real example. What is the maximum current that we can pass through a $1_{2}$ watt, 47 ohm resistor?

$$
\begin{aligned}
& I=\sqrt{ }(P / R)=\sqrt{ }\left(1_{2} / 47\right)= \\
& \text { just under } 0 \cdot 01 \mathrm{~A}(10 \mathrm{~mA})
\end{aligned}
$$

In transistor and other semiconductor circuits, currents are usually very low, rarely rising over a few tens of milliamps. In these cases we will rarely find resistors over $1_{2} \mathrm{~W}$ and usually not more than ${ }_{4}^{1} \mathrm{~W}$. It is only where large currents are flowing (as in power supplies or the output stages of amplifiers) or high voltages are present (as in valve circuits) that we encounter high wattage resistors.

## COLOUR CODING

Resistors are usually marked with their values using three coloured stripes on the body of
the resistor. The first two indicate the two digits in the value and the third the multiplier. Thus, for instance, red red orange is 22 followed by three noughts, which implies 22,000 ohms.

A fourth band is used to indicate the tolerance of the resistor.

The colour code is summarised in Table 2.2.

## TABLE 2.2

## RESISTOR COLOUR CODE

Carbon and metal oxide resistors normally have their ohmic value printed on the body in some form of colour code taking the form of four coloured bands. Values are evaluated with the use of the table below:

| Colour | 1st/2nd <br> digits <br> (A/B) | Multi- <br> plier <br> (C) | Toler- <br> ance <br> (D) $\pm \%$ |
| :--- | :---: | :---: | :---: |
| Black | 0 | 1 | - |
| Brown | 1 | 10 | 1 |
| Red | 2 | $10^{2}$ | 2 |
| Orange | 3 | $10^{3}$ | 3 |
| Yellow | 4 | $10^{4}$ | 4 |
| Green | 5 | $10^{5}$ | - |
| Blue | 6 | $10^{6}$ | - |
| Violet | 7 | $10^{7}$ | - |
| Grey | 8 | $10^{8}$ | - |
| White | 9 | $10^{9}$ | - |
| Gold | - | $10-1$ | 5 |
| Silver | - | $10-2$ | 10 |

EXAMPLE: A resistor colour coded as Orange-white-red-silver, would have a value of $3 \cdot 9 \mathrm{k} \Omega \pm 10 \%$.


Fig. 2.4. The construction of a typical potentiometer.

## POTENTIOMETERS

In electronic circuits the requirement often arises to be able to change a certain parameter (volume, brightness, tone, etc.) under manual control. The cheapest and most readily available variable component is the variable resistor or, in its usual form, the potentiometer.

A potentiometer is a threeterminal deyice and has quite a simple internal construction (Fig. 2.4). It consists of a resistive track either of carbon or similar material (though sometimes it is a coil of wire) with electrical contacts at either end brought out to two terminals. Electrical connection is also made to a third terminal but this can make contact anywhere along the track, the actual position being set manually either by a rotating shaft to which the wiper is mechanically but not electrically connected or, in the case of slider potentiometers, by a linear movement.

To use the potentiometer as a variable resistor, the movable contact and one of the other terminals are used. With the wiper at one end of the track there will be virtually zero resistance between the two terminals; with it at the other end, the full resistance of the track will be seen. At intermediate positions a resistance dependent on the position of the wiper will be seen (Fig. 2.5a).

The most commonly used type of potentiometer has a linear relationship between wiper movement and resistance. In other words if wiper movement is plotted against resistance a straight line is seen. However, the need sometimes arises for a potentiometer with a non-linear characteristic. The most


Fig. 2.5 (a) A linear potentiometer has a linear relationship between the wiper position and resistance whilst (b) a logartithmic potentiometer produces a non-linear graph.
common type of this sort is the logarithmic type. The relationship between the wiper position and resistance being shown in Fig. 2.5b.

Such potentiometers are used where the parameter to be varied does not have a linear relationship to any easily varied circuit para. meter. For instance, sound output power is a logarithmic function of electrical power so varying electrical power with a linear potentiometer would give large changes in volume at one end of the potentiometer and small changes at the other. Using a logarithmic potentiometer evens out the adjustment over the range of the potentiometer.

## MEASUREMENTS USING POTENTIOMETERS

The name "potentiometer" sometimes gives rise to confusion as it does not appear to be any sort of "meter". However, with suitable calibration and the use of the simplest of meters it can indeed be used for measuring.

If a voltage is placed across the resistive track then the wiper of the potentiometer can be used to tap off a proportion of this voltage (Experiment 2.1). If a simple meter is placed in the wiper of the potentiometer then it will indicate when current flows out of or into that wiper.

An unknown voltage (which must be less than that across the potentiometer track) can now be measured by connecting it across
the wiper and one end of the potentiometer. Providing the knob is calibrated we can simply adjust the wiper until no current flows and this can only occur when the unknown voltage exactly equals that across the potentiometer.

## LIGHT DEPENDENT RESISTORS

Ordinary resistors are designed so that external influences such as light, heat and mechanical stress have very .little effect on the nominal resistance. There are, however, special resistors which exhibit marked changes in resistance with these influences.
Light dependent resistors (l.d.r.s) are made of a special material which produces more conduction electrons when exposed to light. They should not be confused with solar cells which are sources of e.m.f. not completely passive as l.d.r.s are.

Experiment 2.2 shows a simple light meter using a readily available l.d.r.

## THERMISTORS

Another type of resistor called a thermistor exhibits large changes of resistance with temperature. Any heating tends to increase conductivity since electrons get "knocked off" as the heat agitates the atoms. However, in thermistors, the materials are specially chosen so that the changes are large.

## EXPERIMENT 2.1 : A SIMPLE VOLTMETER

## Components needed: $100 \mathrm{k} \Omega$ resistor

To use a potentiometer as a voltmeter, the scale of the potentiometer needs to be calibrated. Because the track is linear, we know that equal divisions on the scale will represent equal changes in resistance. Thus it is simply necessary to divide the scale into ten equal increments using for Instance a protractor.

Note that the rotation of the knob is restricted to 270 degrees (three quarters of a full rotation) so only this part needs to be divided up (see FIg. 2.6(c)). Each of the ten divisions can be further subdivided into two or maybe ten if it is intended to try to make more accurate measurements but since the battery voltage is not known to a high degree of accuracy this is not really a practical proposition.
The circuit of the simple voltmeter is shown in Fig. 2.6(a) and the board layout
in Fig. 2.6(b). Note the $100 \mathrm{k} \Omega$ resistor in series with the meter. This is not really part of the voltmeter but serves to protect the meter should the wiper of the potentiometer be at OV and the positive r end of the meter connected to a voltage.
Each division of the scale represents one tenth of the voltage across the potentiometer; In this case 9V. Connect say a
1.5 V torch battery across the "voltmeter" terminals (note the polarity). Adjust the potentiometer until the meter reads zero, that is mid-scale. Read off the scale.

The reading should be about 1.7 corresponding to a voltage of approximately 1.5 V . Note that the "meter" cannot read more than the voltage across the potentio-meter-in our case 9V.

$$
\operatorname{mon} \operatorname{cosin} \theta
$$

Fig. 2.6. A simple voltmeter which can be built on the TutorDeck. (a) shows the circuit diagram and (b) the component layout on the deck. (c) shows the potentiometer scale.
 -

$\qquad$




Fig. 2.9. Diagram of the Eurobreadboard indicating how Individual sockets are permanently interconnected inside the board.

Fig 2.6c (below).



©


Fig. 2.8. Left and right hand panels of the Tutor Deck.

Since current through any resistive element tends to produce heat, these resistors tend to exhibit a resistance which goes down as current goes up.

In older types of television receivers one could find thermistors in the heater circuits of the valves. These valve heaters tend to have very low resistance when cold so a

## PART 2 QUESTIONS

2.1. A resistor of 100 ohms has 5 mA flowing through it. What is the voltage across it:
a) 0.6 volts
b) 5 volts
c) 0.05 voits?
2.2. 250 volts is applied across a $10,000 \mathrm{ohm}$ resistor. How much current will flow:
a) 2.5 amps
b) 25 mA
c) 250 mA ?
2.3. How much power Is disslpated by the resistor in the previous question:
a) 8.25 watts
b) 0.625 watts
c) 25 watts?
2.4. What value is a resistor with the colour code yellow, violet, red:
a) 47 ohms
b) 4700 ohms
c) 270,000 ohms?
2.5. What colour code will a resistor of 150,000 ohms have:
a) brown, green, yellow
b) brown, green, orange
c) yellow, green, black?

## PART 1 ANSWERS

1.1. b) 1.2 . d) 1.3 . c) 1,4 . b) 1.5 .b) and c)
thermistor was used to limit the initial current but to allow the right current to flow once the heaters warmed up.

The thermistor just described has a negative temperature coefficient, this is indicated by the sign $-t^{\circ}$ (see Fig. 2.3c). There are also available positive temperature coefficient ( $4 \mathrm{t}^{\circ}$ ) thermistors. In the case of these devices, their resistance increases when the current increases beyond the "normal" working current.

## STRAIN GAUGES

If a piece of wire is stretched it tends to reduce its cross section which in turn tends to increase its resistance. This principle is used in strain gauges which are used to measure mechanical stress. Thin conductive layers are formed on a flexible substrate. When the substrate bends the conductive layen is stretched and the resistance changes.

In Part 3 we will look at electric circuits and see how Ohm's Law enables us to calculate currents in each component of a circuit


The birth of electronic sound generation was probably around the time of the early valve-operated radios, which succeeded the old "cat's whisker" crystal sets. The use of electronic vacuum tubes, or valves, brought with it the property of amplification, which is the boosting of the minute signals from the radio aerial.

With amplification came the possibility of feeding back a boosted signal in order to further boost the overall result. An adjustable control was provided so as to allow accurate feedback to be set such that the maximum boost would occur, without overdoing it and causing over-feedback which resulted in oscillation.

Over-use of the "reaction control" as it was known, caused all manner of squeaks and whistles to emerge from the then-popular horn loudspeaker! Enter the new age of elec-tronically-produced "music", as the earlier version of the audio oscillator was born.

## ELECTRONIC ORGANS

It was not long before oscillators were used to produce the basic tone generators of the first valve electronic organs. These used a bank of twelve such oscillators, each of which produced one note of the top twelve
notes of the organ keyboard. (Twelve notes comprise one chromatic octave, i.e. including sharps and flats or "black" notes).

The remaining octaves were derived, note for note, by dividing the frequency from each oscillator by a factor of two to produce a note exactly one octave lower. For instance, top $C$ frequency would be divided by two to produce the note $C$ one octave lower.
So the tone generation section was built up to include twelve oscillators and one divider per oscillator for each octave below which the keyboard or keyboards spanned. The oscillators and divider stages were left powered and running at their own particular frequencies continuously, and their various outputs selected as required by the depression of a key or keys on the keyboard. This requires at least one wire per key and often more in some designs.
A basic organ schematic is shown in Fig. 1.1, in which it will be seen that the oscillators feed signals to their respective dividers, from which a large number of individual signals emerge, one for each note of the keyboard (or keyboards). Sometimes these signals are switched direct by the keyboard, but in this example
gating circuits are shown which do the switching electronically, which is more common nowadays.
The signals "chosen" by the depressed keyboard keys are commoned on to a single line in the diagram, but often these are fed out on a separate line per octave for reasons we need not worry ourselves at this point. The selected notes are fed to a Tone Forming circuit. The purpose of this block is to add the desired quality of sound which would be absent were we to listen to the "raw" signals produced by the oscillators and dividers.
In aotual fact, the waveform of the dividers and oscillators is normally a squarewave, which is the shape shown in the diagram. If this shape is amplified and reproduced in a loudspeaker, it is similar to the sound of a clarinet. Obviously, it is not desirable for our organ to sound like a clarinet all the time, or any other single instrument, for that matter. So the squarewave signals are passed to the Tone Forming circuits for modification.

The circuits in this block perform various forms of modification on the signals fed into it. Each circuit is designed to modify a squarewave to produce a more complex waveform which will resemble a different instru-
ment, e.g. trumpet, flute, violin, etc. The circuits are switched in and out by the Stop Tab switches, one stop tab per effect.

The stop tabs may be used singly or collectively to produce a myriad of different effects, and the final composite signal is passed to the output amplifier, via the Swell Pedal for amplification and reproduction as sound by the speaker.

## SYNTHESISER PRINCIPLES

So much for the very basic principles of electronic organs. Now for the very different philosophy of synthesiser design. For the purpose of this series we shall restrict our dealings with the monophonic synthesiser, which is the design which is played one key at a time only. The polyphonic types are currently very expensive and use many of the electronic organ principles.

One of the most striking differences between the electronic organ and the monophonic synthesiser is the latter's comparative simplicity of design; at least so far as a comparison of the schematic diagrams of the two instruments is concerned. The actual circuit design of the component blocks of the synthesiser are by no means simple, as very high accuracy of performance over wide ranges of use must be maintained.

## OSCILLATORS

In the synthesiser we do not use twelve oscillators, running all the time irrespective of whether they are being used at any one time. Instead, we use one or more (generally two
or three) oscillatoors, which are designed to be very versatile. Each oscillator is made to respond to a certain voltage applied to its "voltage control" input.
The frequency, or pitch of signal created by the voltage controlled oscillator is accurately related to the voltage applied. In order for this to be possible, it is necessary for the oscillator to be widely variable, instead of being fixed at one pltch, as is the case with each oscillator in the organ.

## RESISTOR LADDER

In Fig. 1.2, it is shown how the voltage controlled oscillator is controlled by the keyboard. A chain or ladder of resistors is connected in series between the positive and negative terminals of a source of direct current voltage. A current flows through each resistor, and a portion of the total supply voltage appears across the ends of each resistor. If each resistor is the same ohmic value (same resistance value), then the voltage developed across each will be the same.

Suppose the voltage across each resistor were 0.1 volt, then, starting from the bottom resistor, the first junction of resistors would have 0.1 volts on it, the next one up would have 0.2 volts, the next 0.3 volts and so on.

To each junction of resistors is attached one end of a pair of switch contacts operated by one key of the keyboard. The other ends of the contacts are commoned together and taken to the voltage control input line of the voltage controlled oscillator.


Now, if the bottom keyboard switch is operated, the voltage control line of the v.c.o. (voltage controlled oscillator) is connected to the first resistor junction and 0.1 volts is applied. Similarly, the operation of any of the other keyboard switches will result in a different voltage being applied. Hence, for each key, a different voltage, and a different pitch from the v.c.o.

Notice the outputs from the v.c.o. Three different outputs are shown in Fig. 1.2, though in some designs others are possible. The shape of the waveform differs at each output, but its pitch or frequency does not.
The pitch of all three outputs depends, as stated earlier, upon the voltage applied at the v.c.o. input, but the shape, or tonal quality of the three outputs are different.
The smooth-looking shape at the top output gives a mellow tone, and its shape is known as sinewave. The second output shape, known as triangular, gives a less smooth sound, as may well be expected from its appearance, and is similar, but not identical to the effect on organs known as "diapason". The third output shape is a square-wave, and, as we have mentioned before, this sounds rather like a clarinet.
Already, another difference has appeared between organs and synthesisers; we do not derive all our effects from a single wave-shape, but can have three or more at our disposal, at root, i.e. direct from the oscillators. This is not to say that we do not use any form of tone forming circuits in a synthesiser, but simply that we start with a wider base on which to create our various effects.



We will leave the tone-generation section, as the oscillators are known, at this point, and return to it in more detail later, as there are other important sections which should be introduced to give a wider view on basic principals.

## ENVELOPE SHAPING

Even if we are not musicians, we are able to distinguish one instrument from another, even if the same note is played on each. Why?

Well, already we have touched upon differences in quality of tone, or waveshape. This is only one way by which sounds are distinguished. Another way is the way in which the note commences, sustains, and dies away or decays. These qualities are collectively known as the envelope of a sound.

Consider first, the sound of a piano note. As the internal hammer strikes the strings (there are more than one per note, each tuned to the same pitch), the sound commences almost explosively, and decays away gradually if the key is held down or the sustain pedal is pressed as depicted in Fig. 1.3a. But throughout the length of the audible note period the same pitch is created. The volume or amplitude of this pitch, however, starts large, and diminishes with time. If, on the other hand, the piano key is struck and immediately released, a damper is applied to the strings and the note starts abruptly as before, but ends almost as suddenly as shown in Fig. 1.3b.

Already, we have met two different shapes of envelope. One has an abrupt beginning or attack, and a slow decay, and the second has abrupt attack again, but also abrupt decay.
A third example, for good measure, would be the bowed note of a violin.


If the player draws the bow slowly and gently over a string, gradually pressing the bow harder over its travel, the note will build up attack slowly, and give a long attack period. When the bow is removed, the string will slowly decrease its vibrations and a long decay will result (Fig. 1.3c), as in the sustained piano note considered. Notice that the envelopes do, in fact, envelope the waveforms of the three examples, and hence the name.

## ENVELOPE GENERATION

In synthesisers, we produce envelopes, as with other effects, electronically. This involves the use of special circuits which have variable parameters with respect to time. We will consider this in more detail later.
In order that the envelope shaper circuit can perform its task, it must be informed when it is to do so. The instant that a key is pressed on the keyboard, a signal is sent to the envelope shaper to tell it a note is being played. The envelope shaper will have built into it the controls required to set the attack and decay rates. When a key is pressed, the attack of the envelope will be commenced from this instant. If a long attack is required, the signal from the v.c.o. will be gradually allowed to pass through the envelope shaper with increasing amplitude until full strength or volume is reached. If short attack is set, the full signal will be passed immediately through the envelope shaper.

But what about decay? Attack is easy, as we have just seen, but if we press a key in Fig. 1.2 and release it, we see that immediately the release occurs the contacts of the key separate and the voltage on the v.c.o. input line disappears! So, with the
best envelope shaper in the world, if there is no signal to apply a decay shipe to, we cannot shape it.

What we need is some way of telling the v,c.o. to stay oscillating after any key is released, and to remain sounding that note for some time afterwards, but to change its pitch immediately any other note is pressed. This circuit is not an unduly complicated device, thanks primarily to the facilities offered by the v.c.o. design. The circuit, known by function as pitch memory is called in electronic terms a "sample and hold" circuit. It is placed electronically between the keyboard pitch selection line and the input of the v.c.o., and its basic function is to use a capacitor which charges up to the voltage selected by a keyboard switch. When the switch is released, the capacitor charge remains, and, via a special circuit, holds the v.c.o. input line at the same voltage until it is "told" to change to a new value by the depression of another key.

## PORTAMENTO

A useful spin-off from the use of the sample and hold circuit is the simple inclusion of another valuable function, known in musical terms as Portamento. When portamento is applied, instead of the pitch memory changing the voltage at the v.c.o. from one value to another as a new note is pressed, the change is made variable in velocity, i.e. the note will "glide" from the last note played to the next played.

Fig. 1.4 shows a schematic of all the facilities discussed so far. The envelope shaper is triggered simultaneously with the application of a voltage to the pitch memory, by means of a second contact on each key of the keyboard. These contacts
are known as the envelope control contacts. In Fig. 1.4 they are connected to the positive voltage line and are all commoned at each end, so that operation of any one will connect the envelope control line to the positive rail, telling the envelope shaper circuit when to start shaping, and when to start decaying the signal.

Other refinements can be incorporated into the envelope shaper, such that the decay can start before a key is released, but the same basic principle applies.

## FURTHER COMPARISONS

Having considered the basic circuits in a synthesiser, a further comparison with electronic organs would not be out of place. Our simple organ circuit did not consider envelope shaping. This is because few organ manufacturers find it economical to provide very much in the way of shaping.

Sustain is often supplied, but in a conventional organ design, this means providing a separate decay circuit for every note of the keyboard! Admittedly, the circuit is not as complex as our envelope shaper in the synthesiser, but it must be provided in bulk!

Again, attack can be provided in organs, but where provided it is generally either present or absent, as set by a switch, and attack is normally restricted to a very short relative time.

Portamento on organs is rare or non-existent. Sometimes a "glide" facility is provided, which gives a smooth flattening of the played music, of at best about a semitone. Portamento in a simple synthesiser can be applied simply by making the pitah memory capacitor charge slowly through a variable resistor!

Another feature offered by most organ manufacturers is vibrato. This
is the continual variation in pitch of all notes, and is achieved in organs by applying a relatively slow sinewave to each oscillator to change its pitch up and down alternately by about half a semitone each way. In the synthesiser this is achieved in much the same way by applying alow-frequency sinewave to the keyboard resistor ladder such that it is varied or "wavered" up and down by a small amount. In fact, it may be made more than a small amount if desired, so as to give special effects.

In short, the use of oscillators which are voltage controlled allows many things to be done. As will be seen later, oscillators are not the only circuits which can be voltage controlled, and the use of this principle in synthesisers has created the tremendous versatility which we associate with them.

To be continued


## Helping Hand

There is no hobby that I am aware of, that is in any way comparable to Electronics, in the possibilities it offers, for developing from a pastime into a truly worthwhile career. The model train enthusiast does not want to be an engine driver, the amateur sailor, a ship's Captain, or the stamp collector wish to run a sub-post office, now with your electronics enthusiast, I was about to say, "The Sky is the Limit", but with news of America's Pioneer II after a voyage of six years, sending back to Earth pictures of Saturn, would anybody blame me for saying of the electronics enthusiast "His aspirations are bounded only by the Universe'"? I think not !

It is satisfying to feel you are part of the picture and when you reach my number of years you can remember serving young lads with components, and in due course serving their children with similar things. Mind you, it can have its humiliations.

1 remember a young lad (no names, no pack drill) that I served with electronic bits and pieces and now he owns a company with a two million pound turnover and along he comes and offers to buy me outl To think, twenty years ago, I was patting him on the head and complimenting him on his
intelligence. That's where I went wrong, I should have patted him on the head with a brick!

Seriously though, in reality I get a great kick out of every success story, especially if I have played some minor part in helping these novices along the path to success.

## Trouble Shooter

Take for example the case of John Morgan, who used to work for me many years ago. John was undoubtedly a very bright lad and when he emigrated to America his electronic talents were soon spotted. He finished up as chief service engineer (or trouble shooter as they call them over there) to one of the biggest computer companies, at an astronomical salary.

We exchange magazines and occasional letters and he has an Uncle in this country who tells me of his various exploits. Apparently he is so highly thought of that when all else fails they say "Send for John Morgan" and he has a special card enabling him to travel anywhere in the world by whatever mode of travel is the quickest.

Only recently a large engineering firm came to a grinding halt because of a computer failure. The firm was large enough to have four resident engineers but after a three days
struggle they gave up, and the management said "Send for John Morgan". John hops on a Jumbo, a car waiting at the airport whisks him to the factory and twenty minutes later all is humming again. The only people who were upset, were the four disgruntled engineers, who said to John, "Look old man, you might have at least hung it out for half a day or so" !

Well, this country needs all the John Morgans it can produce. A good electronic designer or service engineer will never be without work but this brings me to my final point.

## Next Question

I am often asked why we have no technical staff in our shop and part of the answer is in the difficulty in recruitment. I was forcibly reminded of this the other day when a colleague of mine told me he was trying to find a good knowledgeable lad for his establish. ment. A reasonably large number turned up, some had even completed one year of a City and Guilds course.

To sort out the wheat from the chaff he decided on a few simple questions. Some of the answers were to say the least surprising. One applicant was asked, "What is the purpose of a transformer?" After five minutes deep thought he said "Doesn't it transform Electricity into Copper?" The next question was "If you have an amplifler with an 8 ohm output and four 8 ohm speakers, how would you connect them up?" A long time elapsed and then the lad looked up hopefully and said "With wire?"

Finally, one was asked, "If you connect two capacitors in series, each one, 2 microfarad 1000 volts working, what would be the capacity and the working voltage?" Back came the incredible answer, "The total working voltage would be one, and the capacity 47 ohmsII"-What a pity they had not taken Everyday Electronics regularly!


Thehere have in the past been a number of articles published in the model and home electronics press on the subject of radio control, these have always tended to be either parts of the system or ideas on which a constructor can base a system. These systems have then suffered a further disadvantage in that they are not usually suitable for model aircraft.
What is to be described during the next few months is a radio control system of up to seven channels complete with all the necessary trimmings, which will be capable of being used in aircraft, cars and boats to name the three basic sides to R.C. modelling. Technically the system should be comparable with, and in some cases should be superior to, anything available on the market both in kit and ready-built form and therefore if constructed correctly should

## WINNER!

A new British record was set by Lawrence Armstrong, one of the co-authors of this new series, in the Isle of Man Soaring Championships last August.
Using the prototype EE Radio Control equipment, he kept a model glider aloft for 7 hours 8 minutes, adding $1 \frac{1}{2}$ hours to the old record.
Our author went on to acquire further distinction by securing second place in the Thermal Soaring Competition.
Congratulations Lawrence. You have demonstrated what can be achieved with the EE Radio Control System. Other R/C enthus. iasts will be spurred to reach similar heights using this proven equipment.

The EE Radio Control System is constructed mechanically around parts which are commercially made for the R.C. industry and are also readily available to the home constructor. Electrically the system is constructed on printed circuit board and makes use where possible of integrated circuits to make construction as "fool-proof" as possible. All these components should be available from sources advertising in this magazine.

The equipment comprises the following units:

## Transmitter

## Receiver

## Servos

## Speed Controller

Field Strength Meter

## Battery Charger

Total cost for entire system: $£ 170$ approx. A comparable commercial equipment would cost $£ 225$ plus.

## LICENCE

Before going on any further the constructor should be made aware of the law concerning the use of radiocontrol equipment. As in the case of
all transmitting apparatus a licence is required before the equipment can be used, this can be obtained from: The Home Office, Radio Regulatory Dept., Waterloo Bridge House, Waterloo Bridge, London S.E.1, and costs $£ 2 \cdot 80$ for five years which at 56 p a year is cheap at twice the price!

## SYSTEM CONCEPT

When designing an R.C. system there are many considerations to make, especialy concerning the transmitter, as to the type of circuit to be used. Amplitude or frequency modulation (a.m. or f.m.) for instance. In this case a.m. was chosen because of its longer development and "track" record.

For radio control purposes f.m. is still very young and has not as yet, in the opinion of the authors, lived up to manufacturers' claims in terms



The receiver is a double-tuned-input superhet using plug-in crystals with an i.f. of 455 kHz .

The servos and speed controller use the latest i.c.s.

## THE TRANSMITTER

The complete circuit of the trans. mitter appears in Fig. 1.3. It will be seen that this is composed of four sections: Channel Switching, Encoder, R.F. Stage and Power Supply.

## CHOICE OF ENCODER

The object of the encoder is to 100 per cent modulate the r.f. circuit with a series of pulse widths varying from


Fig. 1.1. A simple half-shot circuit. This works on a CR charging curve where the charging time is determined by the stick pot position. This circuit would be repeated for each channel, and require the setting up of seven pre-set pots.
1 ms to 2 ms dependent upon the position of the sticks on the transmitter.

In starting to design the encoder many things were taken into consideration and it was decided to make the encoder as versatile as possible. Two functions were considered vital: (i) the ability to easily reverse the effect of stick movement on the pulse width, for example increasing instead of decreasing pulse width when the

stick is moved in one direction; (ii) the ability to easily reduce the pulse width variation with stick movement. Although this second feature was not put on the prototype details are given on how to facilitate the feature.

This second consideration is very useful when learning how to fly because a novice always tends to oversteer at first which always ends up in the initial and usually expensive crash. Another useful use for the reduced throw is in cars and boats where during a race a minimum amount of movement is required to complete a course at speed, yet at slow speed a lot of movement is required to manœuvre around.
Most existing commercial systems use a multivibrator driving a series of half-shots the pulse widths of which are controlled by the stick positions. This type of encoder is very difficult if not impossible to arrange
such that the two main facilities now required can be incorporated. The half-shot method is also vulnerable to temperature and supply voltage changes and is also non-linear due to it relying upon a $C R$ charging curve, the curve being its disadvantage.

With the advent of cheap integrated circuits it now becomes possible to design a very versatile encoder which will now be described in detail.

## LINEAR RAMP ENCODER

Fig. 1.2 is a schematic diagram of a linear ramp encoder. This is a simplified version of the final circuit (Fig. 1•3) and uses identical component references. The eight-position switch $S$ however is in reality an electronic device (ICl) as explained later. This switch scans around the potentiometers attached to the control sticks, remaining at each position until the pulse is complete. IC3a forms an

inverting buffer amplifier between these potentiometers and the comparator IC3c.

The capacitor C 6 is allowed to charge up from the constant current source $I$ until the voltage is the same as that at the output of IC3a which in turn, as explained, is dependent upon the stick position. This voltage is detected by IC3c and inverted by IC3d causing TR2 to turn on and discharge C6.

Once the voltage on the capacitor drops below the output of IC3a, TR2 is turned off and C6 allowed to charge up again. The time delay through IC3c and IC3d is long enough to ensure that C6 is fully discharged before TR2 turns off. The capacitor theref $\$$ re is constantly discharged and allowed to charge up to a voltage dependent upon the stick position: thus as this voltage varies so the voltage to which C6 charges varies and as a result the


Fig. 1.2. Linear Ramp Encoder: basic circuit.
time between discharge pulses varies.
Each time $\mathbf{C 6}$ is discharged the switch $S$ is caused to step on to the

next position. It can be seen therefore that the time between successive discharges will depend upon the voltage on each successively selected stick potentiometer, thus producing a series of pulses the widths of which are governed by all the control stick positions in sequence.

## SYNCHRONISING PULSE

In order to synchronise the receiver (described later) it is necessary to have a long pulse between each set of control pulses. This is produced by arranging an eighth position to $S$ which switches in a voltage such that the output of IC3a goes very high causing the capacitor C 6 to charge to a much higher voltage, so producing a much longer pulse than the normal control pulses.

IC3b detects when C6 is discharged and produces a narrow pulse at its output. This pulse is used to both sequence $S$ and drive the r.f. modulator to produce a correctly coded radio signal.

## CAPACITOR TYPE

In practice the type of capacitor used as C6 was found to have a great deal of effect on the circuit performance. After looking at a variety of types, both electrolytic and non-polarised, the best performance was found to be from polyester capacitors, so for best effect a capacitor of this type should be used.

Fig. 1.4 shows the waveforms to be expected at various points in the encoder.
Refer to Fig. 1.3 for the final practical circuit of the encoder.

## ELECTRONIC SWITCH

The switch used to look at each voltage in turn is a cmos analogue switch ICl. This is a device which is dependent upon the digital binary code appearing on pins 9,10 and 11 , will present the signal appearing on one of the inputs on to the output "A" (pin 3) with an effective resistance of 200 ohms .

The code appearing on pins 9,10 and 11 is changed by the counter in IC2 being clocked as alneady mentioned by the output of IC3b so as to present the next channels in sequence on to the output.

VRI-VR6 represent the six stick potentiometers whilst R1, R2 and R3 form the resistive network required for the switch channel (SI). Rll is the resistor used to set the sync pulse width wider than the remaining channel pulses.

## CONSTANT CURRENT SOURCE

The capacitor C 6 is charged from the constant current source formed by TR2, TR4, R21, R22, VR7, D2 and C7. The reference Zener diode D2 is an accurate voltage source over wide temperature and current variations and forms the heart of the current source. TR4 is used purely to cancel out any effects caused by the Vbe of TR2. VR7 varies the current to enable the centre pulse width to be set up on all channels.

## STICK POTENTIOMETERS

As mentioned previously one requirement of the system is to be able to change round the potentiometers on the sticks without affecting the neutral position. This is achieved by arranging that the pot. wiper is in the centre of the pot. when the stick is in the neutral position, thus causing no change in the voltage on the wiper of the potentiometer when the connections are reversed and therefore maintaining the same neutral pulse width whichever way round the pot. is connected.
The second requirement was to be able to reduce the effect of the stick movement on the pulse width. The change in pulse width with stick position is goverend by the gain of IC3a. The gain is the ratio of R14 to whichever input resistor ( $\mathrm{R} 4 \mathrm{R10}$ ) is selected by IC1. It can be seen therefore that the effective pulse width change with stick movement can be altered by changing the appropriate


input resistor. To ensure no change in the neutral position if the gain is changed a biasing network R12 and R13 ensures that when the pot. is in the neutral position the output of IC3a is at the same potential as the pot. wiper.

## REGULATED SUPPLY

To ensure the accuracy of the voltage seen at the wiper of the stick pots the sticks have to be set across an accurate voltage supply. This is achieved by the shunt regulator DI, R17 and C3. Again this uses an accurate Zener reference D1 to maintain a good performance over temperature and supply voltage changes. This regulated supply is also for the reference voltages on IC3a and 1C3b.

Because of the possibilities of r.f, being picked up on the encoder there is a buffer stage made up of TRl, LI and R18 to block any stray r.f. Point " $C$ " then becomes the output to the modulator section.

## MODULATOR

TR6 is the modulator transistor which is used to 100 per cent modulate the P.A. stage; it is driven by the signal "C" from the encoder section. C9 slows down the edges of the modulation envelope thus reducing spurious radiation caused by sharp switching of r.f. signals.

## R.F. SECTION

The requirement of the r.f. stage is to produce a stable 27 MHz signal capable of operating on 25 kHz spacing between channels with as little as possible (and preferably none at all)
radiating interference on other r.f, bands. This r.f. signal then needs to be modulated with the relevant encoded information from the encoder section.

The stable 27 MHz signal is produced by the crystal oscillator TR5, R24, R25, R26, C8 and L2. The output of the oscillator is then tuned by VCl and L3. This series-tuned circuit serves a second function in tuning the input of the power amplifier TR7 and so making for a more efficient stage.

The power amplifier TR7 is a standard Class C r.f. amplifier with L4 as a collector load. R31 is introduced to reduce the $Q$ or "goodness" of the load L4, thus avoiding any instability in the P.A. stage.

## TUNED OUTPUT

The $T$ network of the P.A. stage formed by L5, L6 and VC2 serves two purposes. First it enables the output impedance of the P.A. stage to be matched to the impedance of the aerial in use; second it filters out any harmonics which may be present in the r.f. signal. Clo is introduced to provide a d.c. block to the aerial to avoid excessive d.c. currents flowing should the aerial become accidentally shorted to the transmitter case or even ground, for instance when the transmitter is left switched on on damp grass.

Fig. 1.5 shows the relationship of the modulation envelope to the incoming encoded signal " $C$ ".

## POWER SUPPLY

The whole of the transmitter circuits run off a 9.6 V nominal voltage battery supply. To enable the state of

## Inductors

$\left.\begin{array}{ll}\text { L1 } & 3 \cdot 3 \mu \mathrm{H} \text { r.f. choke } \\ \text { L2 } & 10 \mu \mathrm{H} \text { r.f. choke } \\ \text { L3 } & 3 \cdot 3 \mu \mathrm{H} \text { r.f. choke } \\ \text { L4 } & 10 \mu \mathrm{H} \text { r.f. choke } \\ \text { L5 } & 6 \cdot 8 \mu \mathrm{H} \text { r.f. choke } \\ \text { L6 } & 10 \mu H \text { r.f. choke }\end{array}\right\}$ Maplin

Battery
B1 9.6 V 500 m AH button cell Nicad battery pack

## Meter

ME1 miniature meter $100 \mu \mathrm{~A}$ d.c. f.s.d.
Sockets
SK1 p.c.b. socket block 3-pin 7-way with plugs (SLM)
SK2 DIN socket 3-way
SK3 crystal socket, horizontal mounting (SLM)

## Miscellaneous*

Nicad button cell end-caps (2 off)
Dual-axis open gimble sticks including 2 potentiometers (2 off)
Single-axis auxiliary sticks including
1 potentiometer (2 off)
Metal case with plastics side panels.
Aerial, Aerial base.
-All available from SLM Model Engineers, Cheltenham.
these batteries to be monitored a small meter MEI is used to measure the supply voltage. The batteries used are the nickel-cadmium type of rechargeable cells and as such have a very shallow discharge curve during their "useful life" after which the voltage drops off very quickly.

A fully charged eight-cell pack gives around 10 volts out, and fully discharged 8.5 volts-so to enable us to see this discharge process in more detail we can use an offset meter technique by inserting a Zener diode (D3) in series with the meter which then gives the meter a starting voltage of 8.2 V in the low position. R23 is then used to set the full-scale voltage. With a $100 \mu \mathrm{~A}$ f.s.d. meter a 15 kilohm resistor gives full scale of around 10 V .

It will be found that after the batteries have been taken off charge and the set switched on the meter needle will probably hit the end stop; however it will soon settle down away from the stop after a couple of minutes use.

| TABLE 1.1. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| CHANNEL CONFIGURATIONS AVAILABLE AND REQUIRED STICKS |  |  |  |  |
| No. of Channels | SIngle Axis Sticks | $\begin{aligned} & \text { Dual } \\ & \text { Axis } \\ & \text { Sticks } \end{aligned}$ | Aux Stick | Switch |
| 2 | 2 | - | - | - |
| 3 | 1 | 1 | - | - |
| 3 | 2 | - | 1 | - |
| 4 | - | 2 | - | - |
| 4 | 1 | 1 | 1 | - |
| 5 | - | 2 | 1 | - |
| 6 | - | 2 | 2 | - |
| 6 | - | 2 | 1 | 1 |
| 7 | - | 2 | 2 | 1 |

TABLE 1.1.
CHANNEL CONFIGURATIONS AVAILABLE AND REQUIRED STICKS

## HOW IT WORKS

The EE Radio Control System is a pulse proportional system utilising the 27 MHz radio band. Like all forms of remote control the idea is to transmit information from one place to another in order to control some function and in this case control a model, whether it is a car, boat or alrcraft.


The information starts out as a voltalde across a potentiometer connected to the control sticks. This voltage is thefi converted into a digital pulse whose width is proportional to the voltage. Septeral of these pulses are grouped together into a series pulse train, one for eadifunction to be controlled, and the whole train is repeated 50 times each segfond to enable changes in information to be quickly transferred to the model

With the Informatipf now in digital form, it is then transmitted by the radio waves to the receiveftby the swltching on and off of the carrier wave (amplitude Elevator Aeriol modulation). The radio waves are received by the recelver in the same way as a normal domestic receiver and then the pulse train is fed into a decoder where the pulses are split up into their indivldual channels. Each pulse now goes into a servo which converts this variable pulse width into the physical movement of a control arm which can then be used to move a particular control function of the car, boat or aircraft.


Fig. 1.5. Related waveforms of the encoder output and the modulated envelope of the P.A. stage output.

## CHARGING

Charging is accomplished by connecting to pins 2 (earth) and 3 ( +ve ) of the DIN socket, when the set is switched off, and passing a constant current through the cells. More details of this will be given when the charger is described later in the series.
Another facility on the set is to be able to use an external power source by connecting to pins 2 (earth) and 1 ( +ve ) on the DIN connector. This was used by the authors to enable the transmitter to be used for long days on the flying field where the five hours to be expected from the internal batteries was not sufficient. Switch 52 must be set to "off" when using an external power supply, otherwise the internal battery will be "on-charge".

## HOW MANY CHANNELS

The system as already described has seven channels, so the components list shows the components required for all seven channels. However, depending upon your requirements (and pocket) you can in fact build any size of system from two channels up to the full seven channels.

Next month we will be describing how to construct a transmitter covering from two to seven channels. In the intervening period you can make up your mind on your system size and purchase the required parts.

In order to help you Table 1.1 shows some of the many channel configurations available and the required sticks. When deciding upon the system size do not just judge upon your present requirements but try and plan for the future as modifications later on can be very messy and untidy. We ourselves strongly advise the full system as this should see you through a good few years service and give you good value for money.

Next Month: Bullding the transmitter

# LETTERS 

## Great Interest

I am writing this letter to express my thanks to your great magazine (EE). I started to buy EE two years ago; and when I recelved my coples I read them with great interest but deep down I didn't understand a word of the scientific jargon, but within the two years of reading EEI have become famillar with most of the Elactronic Worid including the Microprocessor and I have already built a Labcentre designed to my needs. So I thank you for the knowledge I now possess.

S. Barton,<br>Spalding,<br>Lincs.

## Sound Division

I have built your Sound-to-Light Unit with 3 Channels.
I thought that you may be interested to see how I divided my frequencies; bass, middle and treble, see Fig. 1.

Thank you for a most interesting magazine.
M. A. Garty,

Bristol.

to transformers

## Hot Ferric

I have only just read the excellent articie on making Printed Circuit Boards (January 1979) and while \& cannot fault it, I think a word of warning might not be out of place.

A year or two ago we produced our own Etching Kits and In the process I learnt quite a lot about Ferric Chloride, Judging by the picture in the artcle the Ferric Chloride used by the wrlter is a fairly weak commercial type, rock hard and not too easy to dissolve but it has the advantage of having no heat problems.
There is on the market to-day quite a big quantity of Ex-Government, pure anhydrous Ferric Chloride which is almost a different substance. It is usually double packed in thick plastlc and double sealed. It has the appearance of dark brown ground coffee and it is much stronger. About one and a half desert spoonfuls (plastic of course) would make enough etching solution for several boards. Its one drawback is that it produces intense heat in contact with water. We advise customers aiways to add the crystals to the water a little at a time, and not the other way round.
To give you a rough idea of the heat generated, if you add something less than two desert spoonfuls to a jam jar, one third filled with water, by the time the last of the chemical is added, it is too hot to pick upl Another odd side effect we found, and that is, if you make the solutlon a little too strong, no etching will take placel
Although it is always looked upon as poisonous and corrosive and should always be treated as such, you may be surprised to learn that it was used for water purifica. tion by the American Forces.
A. Sproxton, Director, Home Radio, Mitcham.

## Better Reception

I have just completed the construction of your Pocket Radio, shown in the June 1979 issue. I have found the performance was very poor, the volume control having ilttle effect on the volume being produced. I narrowed this problem down to C4, value $10 \mu \mathrm{~F}$, this takes several seconds to charge up and therefore is too large. I replaced it with a smaller $0.1 \mu \mathrm{~F}$ non-electrolytic capacitor. This enables the volume control to be used to the best of its ability.

Ilive in an area of strong signal strength, but the radio still gives a poor performance I declded, therefore, to use an external aerlal-a $30 f t$ piece of gash co-ax cable. This can be plugged in to the radio when it Is used in my bedroom. (A 0.1 to $0.22 \mu \mathrm{~F}$ capacitor was placed between the aerial and the tuning capacitor). The Radio now gives a much better reception than before.
I hope this information may prove useful to other readers.
K. P. Holohan

Preston,
Lancs.


NICKEL CADMIUM BATTERY MONITOR (September 1979)

On page 587, column 3, paragraph 1, line 7 should read ... equipment switched Off all should...

## Crossword No. 21-Solution




| CNOS |  | 4020 | 50p | 4050 | $\begin{aligned} & 25 p \\ & 80 p \\ & 30 p \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 4022 | 500 | 4060 |  |
|  |  | 4023 | 13p | 4066 |  |
|  |  | 4024 | 40 p | 406B | 13p |
| 4001 | 13 p | 4025 | 13p | 4069 | 13p |
| 4002 | 130 | 4026 | 900 | 4070 | 13p |
| 4007 | 13p | 4027 | 280 | 4071 | 13 p |
| 4009 | 30p | 4028 | 450 | 4072 | 13p |
| 4011 | 13p | 4029 | 500 | 4081 | 13 p |
| 4012 | 13p | 4040 | 550 | 4093 | 36p |
| 4013. | 28. | 4041 | 55 | 4510 | 600 |
| 4015 | 500 | 4042 | 550 | 4511 | 600 |
| 4016 | 280 | 4043 | 500 | 4518 | 650 |
| 4017 | $47 p$ | 4046 | 900 | 4520 | 60p |
| 4018 | 550 | 4049 | 250 | 4528 | 60p |


| TTL |  | $\begin{aligned} & 7473 \\ & 7474 \end{aligned}$ | $\begin{aligned} & 200 \\ & 220 \end{aligned}$ | $\begin{aligned} & 74141 \\ & 74145 \end{aligned}$ | $\begin{aligned} & 55 p \\ & 550 \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 7400 | 100 | 7475 | 250 | 74148 | 900 |
| 7401 | 109 | 7476 | 200 | 74150 | 55 |
| 7402 | 10. | 7485 | 55 | 74151 | 400 |
| 7404 | 120 | 7486 | 200 | 74154 | $65 p$ |
| 7406 | 220 | 7489 | 1350 | 74157 | 40 p |
| 7408 | 120 | 7490 | 250 | 74164 | 55p |
| 7410 | 100 | 7492 | 30 p | 74165 | 55p |
| 7413 | 22 p | 7493 | 25p | 74170 | 1000 |
| 7414 | 39p | 7494 | $45 p$ | 74174 | 550 |
| 7420 | 120 | 7495 | 350 | 74177 | 50p |
| 7427 | 200 | 7496 | 450 | 74190 | 500 |
| 7430 | 12. | 74121 | 25p | 74191 | 500 |
| 7432 | 18p | 74122 | 350 | 74192 | 50p. |
| 7442 | 38p | 74123 | $38 p$ | 74193 | 50p |
| 7447 | 450 | 74125 | 35p | 74196 | 50, |
| 7448 | 50p | 74126 | 35p | 74197 | 50p |
| 7454 | 120 | 74132 | 450 | 74199 | 900 |
| OPTO |  |  |  |  |  |
| LEO's | 0.1 | Sin. 0 | 2 in | each | $100+$ |
| Red |  | 209 | IL220 | 90 | 7.5p |
| Green |  | 11 TI | IL221 | 13p | 12p |
| Yellow | TIL | 213 | IL223 | 13p | 12p |
| Clips | 3 p | 3 p |  |  |  |
| DISPLAYS |  |  |  |  |  |
| OL704 |  | icc |  | 1300 | 1200 |
| OL707 |  | i CA |  | 1300 | 120p |
| FND500 | 0.5 | in CC |  | 100p | B0p |

## SKTS <br> Low profile by Texas <br> 

8 8pin $\quad 8 \mathrm{p} \quad 18 \mathrm{pln} \quad 14 \mathrm{p} \quad 24 \mathrm{pin} \quad 18 \mathrm{p}$ $\begin{array}{llllll}14 \mathrm{pin} & 100 & 20 \mathrm{oin} & 16 \mathrm{p} & 28 \mathrm{oin} & 22 p \\ 16 \mathrm{uin} & 110 & 22 \mathrm{pin} & 17 p & 40 \text { pin } & 32 \mathrm{p}\end{array}$ 3 lead T018 or TO5 socket. 100 each Soldercon pins: 100:50, 1000:370p

## PCBS

VEROBOARD
Size in. $0.1 \mathrm{in}, 0.15 \mathrm{in}$.
$25 \times 1 \quad 14 \mathrm{p} \quad 14 \mathrm{p} \quad$ Cutter 80 p . $\begin{array}{lll}2.5 \times 3.75 & 450 & 460 \\ 25 & 5 \times 5 & 540\end{array}$
$\begin{array}{lll}2.5 \times 5 & 54 p & 540 \\ 3.75 \times 5 & 640 & 64 p\end{array}$ Pin insertion
$3.75 \times 17 \quad 2050 \quad 1850$
Single sided
Dins per 10040 p 40 p
Top qualiny fibre olass copper board. Single sided. Sire $203 \times 95 \mathrm{~mm}$. 60 p each
Dive pens. 75 p each.
RESISTORS Carbon film resist. ors. Migh stobility

E12 series. 4.7 ohms | low noise 10 M . Any mix |
| ---: | $\begin{array}{llll} & \text { egh } & 100 \uparrow & 1000 \\ 0.25 W & 10 & 0.9 p & 0.8 \rho \\ 0.5 W & 1.5 p & 1.2 p & 1 p\end{array}$ Soeciel developmant packa consisting of 10 of each value from 4.7 ohms to $1 . \mathrm{Meg}$ onm (650 res) 0.5 W £ 7.50. 0.25 W E5.70. METAL FILM RESISTORS

Very high thability, low noise rated ot \%W 1\%. Available from 510 hms to 330 k in E24 serres. Any mix.
${ }_{4}$

LNEAR
 $\begin{array}{lll}\text { LF356 } & 80 \rho & \text { NE531 } \\ \text { LM301AN } 260 & \text { NE555 } \\ & 600 & \text { NE556 }\end{array}$ LM308 600 NE556 600 THIS IS ONLY LM318N 750 NE567 1000 A SELECTION! LM318N 450 RC4 1361000 $\begin{array}{llllll}709 & 350 & \text { LM339 } & 450 & \text { SN } 76477 & 2300 \\ 741 & i 60 & \text { LM } 378 & 2300 & \text { TBA800 } & 700\end{array}$ $\begin{array}{lllllll}741 & i 6 p & \text { LM378 } & 2300 & \text { T8A800 } & 700 \\ 747 & 450 & \text { LM379S } & 4100 & \text { TBA810S } & 1000\end{array}$ $748 \quad 30 \rho \quad$ LM $380 \quad 75 \rho$ TDA1022 6200 71068500 LM3900 500 TLOB1 7107 9000 LM3909 65p TLO84 CA3046 55 LM 3911 100 ZN414 80p CA3080 70 D MC1458 32P ZN425E 390D (

## TRANSISTORS

 $\begin{array}{llllll}A C 127 & 170 & B C 131 & 350 & 2 N 3054 & 500 \\ A C 128 & 160 & 80132 & 350 & 2 N 3055\end{array}$ $\begin{array}{llllll}\text { AC128 } & 160 & 80132 & 350 & 2 N 3055 & 500 \\ \text { AC176 } & 180 & 80139 & 350 & 2 N 3 & 130\end{array}$ $\begin{array}{llllll}A C 176 & 180 & 80139 & 350 & \text { 2N3442 } & 1350 \\ \text { AD161 } & 380 & 80140 & 350 & 2 N 3702 & \end{array}$ $\begin{array}{llllll}\text { AD161 } & 38 \mathrm{p} & 80140 & 350 & 2 N 3702 & 8 \mathrm{p} \\ \text { AD162 } & 388 & \text { 日FY50 } & 150 & 2 N 3703 & 8 \rho\end{array}$ $\begin{array}{lllllll}\text { AD162 } & 38 \mathrm{p} & \text { BFY50 } & 150 & 2 N 3703 & 80 \\ \text { BC107 } & 80 & \text { BFY } 51 & 150 & 2 N 3704 & 9\end{array}$ $\begin{array}{llllll}\text { BC107 } & 80 & \text { BFY51 } & 150 & 2 N 3704 & 80 \\ \text { BC108 } & 80 & \text { BFY52 } & 150 & 2 N 3705 & 90\end{array}$ $\begin{array}{llllll}\text { BC108 } & 80 & \text { BFY52 } & 150 & \text { 2N3705 } & 90 \\ \text { BC108C } & 100 & \text { MJ2955 } & 980 & 2 N 3706 & 90\end{array}$ $\begin{array}{llllll}\text { BC108C } & \text { BC } & \text { MJ2955 } & 980 & \text { 2N3706 } & 9 p\end{array}$ $\begin{array}{llllllll}\text { BC109 } & 80 & \text { MPSAO6 } & 200 & 2 N 3707 & 90 \\ \text { BC109C } & 100 & \text { MPSA56 } & 200 & 2 N 3708 & 90\end{array}$

 $\begin{array}{llllll}\text { BC148 } & 70 & \text { TIP30C } & 700 & 2 N 3820 & 44 \rho\end{array}$ $\begin{array}{lllll}8 C 177 & 140 & \text { TIP31C } & 650 & 2 N 3904 \\ 8 p\end{array}$ $\begin{array}{llllll}\text { BC178 } & 14 \mathrm{p} & \text { TIP32C } & 800 & 2 N 3905 & 8 p\end{array}$ $\begin{array}{lllllll}\text { BC179 } & 14 \mathrm{p} & \text { TIP2955 } & 650 & 2 N 3906 & 80\end{array}$ | $8 C 182$ | 100 | $T 1 P 3055$ | $55 p$ | $2 N 4058$ | 120 |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 8 |  |  |  |  |  |

 $\begin{array}{llllll}8 \mathrm{Cl} 184 & 10 \mathrm{D} & \text { ZTX } 108 & 14 \mathrm{p} & \text { 2N5459 } & 32 \mathrm{p}\end{array}$ $\begin{array}{llllllll}\text { BC184L } & 10 \mathrm{p} & 2 T \times 300 & 160 & 2 N 5777 & 50 p\end{array}$ $\begin{array}{ll}\text { BC212 } & 10 p \\ \text { BC212L } & 100\end{array}$ BC2124
BC214
100 BC214L

## DIODES

$\begin{array}{llllll}8 C 477 & 190 & 1 N 914 & 30 & 1 N 4006 & 6 p\end{array}$ $\begin{array}{llllll}8 C 478 & 190 & \text { 1N4001 } & 4 p & \text { 1N5401 } & 13 p \\ 8 C 548 & 100 & \text { 1N4002 } & 4 p & \text { BZY88ser. } 8 p\end{array}$ BC548 10p 1 N4002 $4 p$ BZY88 ser. $8 p$ 8CY71 IN4148-£1.40/100. £11/1000

## CAPACITORS

TANTALUM BEAD
$0.1,0.15,0.22,0.33,0.47,0.68$,
$182.2 \mathrm{uF} @ 35 \mathrm{~V}$
$4.76810 \mathrm{~F} @ 25 \mathrm{~V}$
22@16V.47@6V, 100@3V
$0.001,0.01,0.022,0.033,0.047$
POLYESTER
Mullara C280 series
$0.01,0.015,0.022,0.033,0.047,0.068,0.1,5 p$
$0.15,0.22$
$0.33,0.47$
70
100
140
1.OUF

170

## CERAMIC

Plare type 50 V . Available in E 12 series from
22 pF to 1000 pF and E 6 series from 1500 pF to
0.047 F . RADIAL LEAD ELECTROLYTIC
$\begin{array}{llllll}63 V & 0.47 & 1.0 & 2.2 & 4.7 & 10\end{array}$ $\qquad$

|  |  |  | 220 |  | $20 p$ |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 25 V | 10 | 22 | 33 | 47 | $5 p$ |
|  | 100 |  |  |  | .80 |
|  | 220 |  | 470 | 100 |  |
|  |  |  | 450 |  |  |
|  |  |  |  |  |  |

## CONNECTORS

JACK PLUGS AND SOCKETS

|  | screene | unscreened socket |  |
| :--- | :---: | :---: | :---: |
| 2.5 mm | $9 p$ | $13 p$ | $7 p$ |
| 3.5 mm | $9 p$ | $14 p$ | $8 p$ |
| Standard | $16 p$ | $30 p$ | $15 p$ |
| Stereo | $23 p$ | $36 p$ | $18 p$ |

OIN PLUGS AND SOCKETS

|  | plug | chassis <br> socket | line <br> socket |
| :--- | ---: | ---: | :---: |
| 2oin | $7 p$ | $7 p$ | $7 p$ |
| 3pin | $11 p$ | $9 p$ | $14 p$ |
| 50 in $180^{\circ}$ | $11 p$ | $10 p$ | $14 p$ |
| 50 in $240^{\circ}$ | $13 p$ | $10 p$ | $16 p$ |

1 mm PLUGS AND SOCKETS
Surtable for low voltage circuits, Red \& black. Plugs: $6 p$ each Sockets: $7 p$ each.
4 mm PLUGS AND SOCKETS
Avallable in blue, black, green, brown, red, white and vellow. Plugs 110 each Sockets: 120 each PHONO PLUGS AND SOCKETS Insulated plug in red or black Screened plug 13p Screened plug
Sinyle socket

STGVFNSON Electronic Components SOLDERING:IRONS
ANTEX X25 (25W) or ANTEX CX (17W)
$390 p$ each
240p each
Reel of solder (39.6M)

## LOUDSPEAKERS

56 mm dia. 8ohms. $70 \mathrm{p} \quad 64 \mathrm{~mm}$ dia. 64 ohms. 75 p 64 mm dia. 8 ohms. $75 \mathrm{p}=70 \mathrm{~mm}$ dia. Bohms. 100 p Magnetic earpiece including 2.5 or 3.5 mm plug. 15 p each Crvstal earpiece including 3.5 mm plug.

## SWITCHES

Subminiature toggle. SPDT 70p. DPDT 80p Standard toggle. SPST 34p. DPDT 48p.

Slide switches (DPDT) miniature or standard 15p. Push to make switch. 15p. Push to break switch. 20p. Wavechange switches: $1 \mathrm{P} 12 \mathrm{~W}, 2 \mathrm{P} 6 \mathrm{~W}, 3 \mathrm{P} 4 \mathrm{~W}, 4 \mathrm{P} 3 \mathrm{~W}, 43 \mathrm{p}$

## CONTROL KNOBS

Ideal for use on mixers etc. Push on type with black base and marked position line. Cap available in red, blue, green, grey, yellow \& black. 14 p.

## MISCELLANEOUS

Connection cable available in single or stranded packs of eight colours. 8 metre pack 40 metre pack Sing
18 p Stranded 40 metre pack 85p

18p
$80 p$
Battery clips for PP3 with lead. $6 p$ each. Battery clips for PP9 with lead. 10 p each.
 Miniature crocodile clips in red or black. $8 p$ each. Red or black probe clips. 20p each. Murata Ultrasonic Transducers. 180p each. 350p pair.

## PANEL METERS

|  |
| :---: | High quality $2^{\prime \prime}$ wide view meters. Zero adjustment. Back illumination wiring. Available in $50 \mathrm{uA}, 100 \mathrm{uA}, 500 \mathrm{uA}$, $1 \mathrm{~mA}, 100 \mathrm{~mA}, 500 \mathrm{~mA}, 1 \mathrm{~A} . £ 4.75$ éa. VU meter similar style. $£ 1.40$ ea.

## SLIDE POTENTIOMETERS

Good quality 60 mm
travel slider with
80 mm fixing centres.
Available from $5 k-500 k$
In $\log$ and linear. 55p each.
Suitable black knobs 6p ea. Coloured knobs $10 p$ ea.
We now offer one of the widest ranges of components at the most competitive prices in the U.K. See catalogue for full details. We welcome callers at our shop in College Rd, Bromley, from Mon-Sat, 9am-6pm (8pm on Weds and Fridays). Special offers always available.
We also provide an express telephone order service. Orders received before 5pm are shipped same day. Contact our sales office now with your requirements. TELEPHONE: 01-464 2951/6770.
Quantity discounts on any mix $\mathrm{THL}, \mathrm{CMOS}$. 74 LS and Linear circurts: $100 \cdot 10 \%, 1000$. 15\%. Prices VAT melusive. Please ard 30 p for carriage. All prices valid to April 1980. Olficial orders welcome.

ORDERS DESPATCHED BY RETURN

BARCLAYCARO \& ACCESS WELCOME

# Everyday News 

## BIG REWARDS FOR MICRO IDEAS

Three announcements this month (September) help to highlight the efforts being made to get to grips with the microelectronics revolution.

## BRITISH MICROPROCESSOR COMPETITION

Suddenly everyone can get into the microprocessor scene, yes even amateurs, by entering the British Microprocessor Competition organised by its Joint sponsors-the National Research Development Corporation (NRDC) and the National Computing Centre Limited (NCC). Their aim-to stimulate and encourage British innovation in the use of microprocessors in any type of product, process or service. This is a competition for the best invention incorporating a program. mable microelectronic device.
Prize money totalling $£ 20,000$ will be awarded to entries with working models, and those without a working model. First, second and third prizes in the working model category are $£ 10,000, £ 5,000$ and $£ 2,000$ respectively, whilst first and second prizes in entries without working examples are £2,000 and $£ 1,000$.

The competition is open to all individual residents in the UK, including UK registered companies, and other organisations located in the UK such as universities, polytechnics and other institutions engaged in education or research.

The NRDC and NCC staff will judge the competition with 4 main criteria in mind -the degree of novelty, its potential commercial value, the technical and commercial viability and the standard of documentation.

Although the winners names will be announced next year their ideas will be protected; publication only taking place when patent protection exists. All rights
are protected for the de signer and there is no obligation for further involvement by either party.

The NRDC, which this year celebrates 30 years of idea development, have indicated their willingness to look at non-winner ideas along with the winners inventions with a view to offering financial support to develop them. A sum of half a million pounds has been allocated to provide just this backup!

The closing date of the competition is Friday, 14 December 1979 and official Entry Forms and details are freely available from The National Computing Centre, Oxford Road, Manchester M1 7ED.

## International Prestel

The British Post Office is to test-market an inter. national Prestel service for travelling businessmen and government officials.

The trial is planned to last a year and will cover selected users in up to six countries. If there is sufficient interest the international service will be additional to the UX national Prestel service.

## YOUNG ENGINEER FINALS

[^0]
## NATIONAL MICROELECTRONICS COMPETITION

A rent-free $£ 30,000$ factory for one year is one of the inducements being offered by the Peterborough Develop ment Corporation in the National Microelectronics Competition.
The aim of the NMC is to find ideas which are simple to manufacture and have got a ready market. Top prize is $£ 4,000$ and the only restriction is that no company with a turnover in excess of £2 million may enter. The chal-
lenge is to prove that the application is technically sound and that it can be produced and sold at a profit.
The Corporation, with the sponsorship of Barclays Bank and Finance for Industry, offers apart from the new factory, the prospect of £250,000 venture capital from Finance for Industry.
Closing date for the National Electronics Competition is 31 January 1980.

## REGIONAL HELP

Another local authority promoting interest in microelectronics is the Lothian Regional Council of Scotland. They plan to fund a micro aid plan to the tune of $£ 350,000$ over the next five years, which they hope will bring microelectronic tech. nology to companies in the area.
This initiative will bring the Edinburgh University Wolfson Microelectric Institute directly into contact with local firms regardless of their level of technical knowhow. They also hope that local schools and poly. technics will become in. volved.
Part of the $£ 70,000$ per year investment will go towards setting up a new professorship of microelectronics at Edinburgh University and also help to fund three high level engineers, who will seek potential applications of microelectronics. The engineers will approach companies rather than wait for potential micro users to make the first response.

## Boss sells Boss

Having built up Boss Industrial Mouldings Ltd., into one of Europe's largest manufacturers of enclosures, indicators, breadboarding systems and other hardware products, Ian Boss has formally sold all his interest in the organisation which now becomes part of the Pistor Elektrotechnik Group of West Germany.

## ANALYSIS

## THE FILLING IN THE SANDWICH

There are many big producers who are not mass producers, but batch producers of many different products. A batch may be half a dozen units or fifty or so. They may be for specific customers with different delivery dates. Individual finished units may need to be married up into a system and tested as such before shipment. The number of different units being made at any one time may run into hundreds.
This is the sort of manufacturing operation undertaken at Hewlett-Packard's minicomputer facility at Grenoble, France. Cyril Yansouni, the plant's general manager, had quite a problem in keeping tabs on where every product in various stages of assembly was and what was happening to it. He already had those two indispensables, computeraided design and computer-aided automatic test equipment at the outer ends but needed, as it were, the filling in the sandwich.
He calls it CAM (Computer-Aided Manufacturing) and spent 30 months designing the equipment and integrating the system in his own plant.

The cornerstone of his CAM system is shop-floor data capture using specially designed easy-to-use computer terminals at every stage of manufacture to provide realtime product tracking information at every stage of production, assembly and testing.
Over 1,200 products a week pass through the production lines. Each is given a traveller card which stays with it at every stage. The terminals have two slots, one for a badge reader which identifles the person using it, the other for the traveller card which carries data about the product, what it is, who has ordered it etc. Date and time of arrival in a department is automatically transferred with the rest of the data to the central computer.
The result of the exercise is that production is speeded up and bottlenecks eliminated. At the same time the cost of components being worked on along the lines has been cut by about $£ 1$ million despite the factory output having doubled in two years.
Nobody is working any harder than they did before. And nobody is losing his job. In fact they are planning to expand the work-force from 500 to 800 people in the coming year.
Part of this increase is due to the data capture terminals which H-P is now marketing. Over a thousand will have been made and shipped to other manufacturers with similar problems by the end of this year.

Brian G. Peck.

## Engineering Famine

Despite relatively high unemployment figures there is a serious shortage of engineering staff. Earlier this year GEC alone had vacancies for $\mathbf{1 , 6 0 0}$ engineers, $\mathbf{1 , 1 0 0}$ techniclans and 800 craftsmen.
Those training now for the electrical and electronics professions and trades need never be out of work.

## VIDEO NEWS

Firm evidence of the growth of electronic news gathering and associated technologies in Europe is provided in the latest contracts placed with Sony Broadcast Ltd.
During the past six weeks orders totalling some $£ 858,000$ have been placed for Sony video recording equipment
by the State Broadcasting organisations of Austria, Italy, Poland and Switzerland.

> A specialist Viewdata Exhibition for information providers and others professionally engaged in using and operating viewdata and teletext systems is to be held at the West Centre Hotel, London, on November 7.8.

## BREADBOARD '79

This year's Breadboard 79, the kits and bits show for the home electronics enthusiast, has moved to larger premises.

The venue is the Royal Horticultural Halls, Elverton Street, Westminster, London, SW1, from 4 December to 8 December inclusive.

Over 90 exhibition stands will feature microcomputer systems, analysers, logic test accessories, hi f amplifier kits, as well as a varied range of construction kits and TV games.

Everyday Electronics will be there.

## MOBILE JAM

Mobile radio channels have become so congested that the Home Office is to conduct trials with single sideband transmission with 5 kHz channel spacing. Present channel spacing with frequency and amplitude modulation is $12 \cdot 5 \mathrm{kHz}$ or 25 kHz .

SSB could double the number of channels usable with no interference, thus allow. ing for considerable expansion of the mobile services used by businessmen and other organisations.

## LOOKING B ACK

A 20 page booklet to mark the 50th anniversary of the formation of Pye Radio Ltd., is now available, free of charge, to readers on application to Pye Ltd., Publications Dept, 137 Ditton Walk, Cambridge.

The Story of Pye Wireless traces the history of Pye Receivers from when they were originally produced by W. G. Pye \& Co. Written by Gordon Bussey the publication is illustrated with photographs of receivers from 1922 onwards and scenes in the Pye factory early years.

## UK-USA PHONE CABLE GETS GREEN LIGHT

The final seal was placed on an international agreement recently for a new 1100 million telephone cable between Britain and the USA that will boost Britain's transatlantic cable links by more than 50 per cent.

At present more than 20 million phone calls are made each year between the UK and USA, and more than half go by cable. The demand for telephone service between the two countries has been growing by a steady $15-20$ per cent a year throughout the 1970s and shows no sign of slackening.

Called TAT 7, this giant submarine system, with a capacity of 4,200 simultaneous connections, will carry phone calls, computer data and telex messages between Europe and the USA and Canada. A sizeable part of its cost will be spent in Britain on cable manufacture.
The new system is due to come into service in 1983. It will run some 3,400 nautical miles between Porthcurno (Land's End) and Tuckerton, New Jersey. At the British end it will continue for some two miles inland, terminating at the Post Office's Land's End repeater station.

The cost of the project is being divided equally between North America and Europe. On the European side, Britain is partnered by 17 other participants and her share-22 per cent of the total, is the largest of all
those. There are seven participants in the project on the North American side, including the American Telephone and Telegraph Company which has the largest single share in the system, amounting to some 40 per cent of the total.

Manufacture of the new system will be shared between the USA, Britain and France. About 2,700 miles of cable will be made in Britain by Standard Telephone and Cables Ltd, under a contract worth some $£ 30$ million.

# SAXON ENIERTAINMENTS <br> P.A. \& DISCOTHEQUE <br> EQUIPMENT AT <br> INCOMPARABLE PRICES 

STANDARD CENTAUR IOOW
£309 incl. of carr Deopsit $\mathbf{£ 6 2 . 0 0}$ 12 months (a) $\mathbf{2 4 \cdot 4 7}$ or 24 months @ $\mathbf{£ 1 4 \cdot 1 9}$

## SUPER CENTAUR 200W

## £366 incl. of carr

Deposit
£74.00
12 months @ $\mathbf{6 2 8 . 9 4}$ or months @ $\mathbf{6 1 6 . 7 8}$

## GXL. 200W With Twin 200 Watl Cabinets

 12 months (8) $\mathbf{6 3 7 . 2 7}$ or 24 months © $\mathbf{4 2 1 . 6 0}$

## GXL WITH PDF BINS (illus.)


£833 incly Of carr Deposit $£ 167 \cdot 00$ 12 months @ 666.03 or 24 months © 838.28


STEREO DISCOS
C/W LIGHT SHOW \& DISPLAY $100 \mathrm{~W}-600 \mathrm{~W}$


XL + PDF BINS t 2 Year warranty Full Mixing + Crossfade + Mic/Tape Inputs

* Headphone \& Cue Light Monitoring * Full Range Bass/Treble Controls+ Mic Tone CREDIT
TERMS TERMS

MINI DISCO 100 WATT MONO SYSTEM WITH LOUDSPEAKERS $£ 229$ incl. of Carr £46.00 12 months @ $£ 18.13$ or 24 months @ $£ 10.52$
P.A. SYSTEMS

2 YEAR GUARANTEE
$£ 207.00$
12 months (3) 816.35 or
24 months © 69.49
太 Four Mixing Inputs
t Twin Piezo Horn Columns
200 WATT \& VAT $\mathbf{1} \mathbf{1} \mathbf{3 0 9}$.00
12 months © $\mathbf{6 2 4} 47$ or $\mathbf{~} \mathbf{~} 62$
AMPLIFIER UNITS ONLY

APIOO AMPLIFIER
 * 4 mixed Inputs - Bass/Treble Controls * 100 W Watts Output

AP200 AMPLIFIER $£ 102.92$ +Carr $£ 1.50$ * Six Mixed Inputs VA Three Sets Bass/Trebl * 200 Watts OustDut ¿ Slave Socket

## JUST PLUG IN AND GO ! !

SEND TODAY FOR YOUR FREE BROCHURE



PLUTO PROJECTORS PI 40 £ 44.27 150 WATT INC WHEEL

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SA308 8 ohms $30 \mathrm{~W} 45 \mathrm{~V} £ 12.36$ Supply for 2 modules $\{13.69$
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## SAXON ENTERTAINMENTS

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[^1]THis piece of equipment has been devised to allow sounds generated in one place to be heard in another. In particular, from one room where a baby or child is situated to another, such as the bedroom or lounge occupied by the parents or baby sitter. The unit uses a microphone to pickup the sounds and these signals are amplified to produce the same sound in a loudspeaker mounted in the control box.

The Baby Alarm is completely safe, and the child is in no danger if he/she "acquires" the unit. The alarm is powered by a single PP3 9 V battery and is economical, quiescent current being approximately 2.5 milliamps.

## CIRCUIT DESCRIPTION

The complete circuit diagram of the Baby Alarm is shown in Fig. 1.

Signals generated in the crystal microphone insert MICl are
passed to a high impedance buffer amplifier TR1, and f.e.t. wired as a source follower. This stage provides no amplification, but is included to provide low loading on the crystal microphone which is essential for a flat frequency response.
The effect of high loading on such a microphone is to provide a very "tinny" effect. Not entirely essential for specified application, this stage does however allow the circuit to be used in other applications where clear speech is required, e.g. an intercom.

The output from the source follower appears across VR1 used as a volume control, and from here to ICl , connected as a noninverting amplifier. Resistors R6 and R7 provide the necessary biasing for an op-amp operating from a single power supply. Gain is approximately equal to the ratio of R5 to R4 i.e. 1000. The output signal is fed to and heard in LSl.


Fig. 1. The circuit diagram of the Baby Alarm including pluggable remote microphone.


## CONSTRUCTION

The prototype unit used a piece of 0.1 inch circuit board size 20 strips x 30 holes. The uppermost five strips are not used electrically but provide space for mounting screws. The prototype used selfadhesive horizontal mounting strips in preference to fixing nuts and bolts.

The layout of the components on the topside of the board and the breaks to be made along the copper strips on the underside are shown in Fig. 2.

Begin by soldering in the wire links followed by the resistors and capacitors. Take care when soldering in the f.e.t.s as these can be easily damaged when being soldered. Use of heatshunts is recommended. F.e.t.s can also suffer damage by "leaky" irons. A couple of turns of tinned wire wrapped around and shorting all leads during soldering will prevent such damage. Remember to remove the wire after: wards. Finally position and solder in ICl.

Sufficient lengths of flying leads should next be connected to the board. A short length of


Fig. 2. The layout of the components on the stripboard and the breaks to be made on the underside of the board; also shown are components mounted to the case and position of battery and circuit board on base panel and full interwiring. Bottom right shows mounting of microphone in case and connection to phono plug via screened cable.

screened lead to connect to the input socket SKl was used in the prototype, but this is not essential.

## CASE

The author used a plastic box to house the unit, approximate dimensions $150 \times 75 \times 45 \mathrm{~mm}$. The case was used "inverted" so as not to show any panel fixing screws. The intended front panel is used for the base panel to which the circuit board and battery are fixed. The latter was secured with a self adhesive foam pad.

Prepare the box to accommodate S1, VR1 and SK1 and drill a pattern of holes above where the speaker is to be positioned to allow the sound to escape and reach the user.

In the prototype the speaker was glued in position using a polystyrene glue. Fix the components and wire up to the board as shown in Fig. 3.

The base panel (lid) can now be secured, and rubber feet fitted for good measure.

## MICROPHONE

The microphone is mounted in a smaller plastic box (inverted as before). Drill a pattern of holes above MICl position and glue the latter in place. Solder sufficient lengths of screened cable to MICI to join the two boxes in their final positions. The cable should pass out through a gripping (or stain relief) grommet and terminate in a plug to match SKI.

## COMPONENTS

Resistors

| R1 | $4 \cdot 7 \mathrm{M} \Omega$ |
| :--- | :--- |
| R2 | $390 \Omega$ |
| R3 | $10 \mathrm{k} \Omega$ |
| R4 | $100 \Omega$ |

R5 $100 \mathrm{k} \Omega$
R6 $10 \mathrm{k} \Omega$
R7 $10 \mathrm{k} \Omega$
All 1 W carbon $\pm 5 \%$

## Capacitors

| C1 | $470 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. |
| :---: | :---: |
| C2 | $0.1 \mu \mathrm{~F}$ plastic or ceramic |
| C3 | $0 \cdot 1 \mu \mathrm{~F}$ plastic or ceramic |
| C4 | $47 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. |
| C5 | $33 \mu \mathrm{~F} 6 \mathrm{~V}$ elect. |
| C6 | $470 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. |
| Semiconductors |  |
| TR1 | 2N3819 n-channel f.e.t. |
| IC1 | 741 differential op-amp |
|  | 8 -pin d.i.l. |

Miscellaneous
$\mathrm{MIC1}$ crystal microphone insert
S1 d.p. on-off rotary switch
VR1 22 kilohm carbon log. law
SK1 phono socket
LS1 miniature loudspeaker $80 \Omega 70 \mathrm{~mm}$ diameter
B1 9V (PP3)
PL1 phono plug
Stripboard: 0.1 inch matrix, 20 strips $\times 30$ holes; PP3 battery connector; knobs (2 off); board mounts; screened cable; grommet; cases (2 off).

## TESTING

Plug the two units together and switch on. A click should be heard in the loudspeaker. Turn up the volume control. If the two boxes are less than about a couple of metres apart, a feedback howl will be heard. With the microphone at a distance from the control box, a sound source such as a portable radio placed near the microphone will be heard in LS1. Turning VR1 clockwise should increase the volume.

Remove the sound source. A small amount of hissing may be heard with VR1 fully advanced. Hum was absent on the prototype. Handling the cable will produce noise; for this reason the cable

should be firmly secured when the unit is finally fitted.

If all is well the units may be fitted in their respective rooms and can either stand on any flat surface or be mounted on the wall, the latter suiting the microphone, keeping it out of reach. A single "keyhole" cutout on the backpanel will allow single screw fixing.

By a suitable switching arrangement, two microphones and two speakers, the Baby Alarm can be converted to function as a two-way intercom.

#  

## SUPER PHOTODETECTOR

It is clear, on looking at prices of photodetectors (photo emissive types, photo transistors, l.d.r.s, etc.) that these devices are by no means cheap; the least expensive component I have found is the 2N5777 photo Darlington at 60 p . With a little care, it is possible to produce one's own photo transistors, at a fraction of the cost.


Fig. 1.
Fig. 2.
Take a transistor in a TO5 or TO18 can, such as a BCl 107 , and, using a fine razor saw carefully remove the top of the transistor, taking care not to squash the can (see Fig. 1). Carefully shake out any particles of metal which may have fallen inside the transistor. You will find that the innards of the transistor are now exposed to the environment, and if light is
allowed to fall onto the chip, you have a photo transistor. If desired, a few drops of cold setting, clear plastic resin may be poured into the can to afford some protection, but this is not essential.
Leaving the base unconnected, in fairly bright sunlight I found that a BCl 107 would pass $200 \mu \mathrm{~A}$. This sensitivity may easily be increased by using another BC107 transistor, the two being connected as a superalpha pair (see Fig. 2). There should now be enough sensitivity to drive a relay without further amplification.
By this method, either $n p n$ or pnp silicon photo transistors can be made, much cheaper than the cost of a ready made device. Also, the response is very fast, better than some l.d.r.s.

Peter F. Vaughan, Lynton.

## BUTTON STOP

When using twin-core (figure of 8) cable, I bind the separated ends of the cable with a small 4holed button. This stops the split in the cable from lengthening, see Fig. 1.
A. A. Moore, Preston, Lancs.


Fig. 1.

Stupendous Offer to Everyday Electromis Readers SOLAR ALARM DUAL - TIME CHRONOGRAPH only

## " 19.95

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TIME ZONE 1
Continuous display of: Hours: Minutes: Seconds or Date
Day.
TIME ZONE 2
Continuous display of: Hours: Minutes: Seconds or Date: Day.
ALARM
Hours and Minutes.
STOP WATCH
Hours: Minutes: Seconds: $1 / 10$ Seconds to 11 Hours 59
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## By Dave Barrington

## Test Case

Knowing the pride constructors take in the appearance of their finished projects, we make no excuse for returning to the subject of cases again this month.
Ideally suited to housing test gear accessories such as signal injectors, logic probes, small counters, voltage and resistance probes, and continuity checkers, the CTP- 1 probe case from Continental Specialties Corporation comes complete with associated hard. ware.


CTP-1 case kit from Continental Specialities.

Based on the case used in their LPK. 1 logic probe kit it is supplied complete with a 3 ft length of two-wire connecting lead with a moulded strain reliever and terminated with "croc clips", a nickel-plated screw-in probe tip, a mating tapped hex probe-tip connector, assembly screws, and a cut to size blank printed circuit board.

Also available from CSC is their latest 32 -page product catalogue which features their range of circuit breadboarding equipment, logic testing devices and test instrumentation.
Products featured include a range of solderless breadboards and bread-
board assemblies, test clips, instrument cases, pulse and function generators, frequency counters and accessories, logic probes, logic monitors and a digital pulser.

Copies of the catalogue and further details of the CTP. 1 probe case. can be obtained from Continental Special. ties Corporation, Dept EE, Shire Hill Industrial Estate, Saffron Walden, Essex, CB11 3AQ.

## Teach-In '80

For those readers about to order components for the EE Tutor Deck and Teach-In 80 experiments, we have just heard that due to increase costs Home Radio have had to increase the price of the complete kits of parts for this project and experiments up to Part 6, to £22•50. (List A-£19. B-£4).
However, we understand that Greenweld and A. Marshall (London) Ltd have no plans, at the present time, to increase their published prices. Also, the following advertisers are able to supply complete kits of parts: Ace Mailtronix, Electrovalue, Magenta and Watford Electronics.

## Tool sets

More renowned for their top grade soldering equipment, Light Soldering Developments Ltd. are now marketing four handy miniature tool sets.

Each set comes in a plastic case with transparent lid and the tools have chromium plated brass handles. The kits are made up of screwdrivers, open and socket spanners and cross. point screwdrivers.

The set of six instrument screw. drivers (Model. 1113), have hardened and tempered steel blades ranging in width from 0.8 to 3.8 mm and retail at $£ 2.93$ including VAT. The 19 piece combination set, type 37228, consists of open and socket spanners, $5 / 64$ in to $5 / 16 \mathrm{in}$ across flats, socket head, cross head and plain screwdrivers, and a scriber and is priced at $£ 5 \cdot 12$.

A set of five metric box spanners, model 37227, with a tommy bar with hardened and tempered steel ends come in a range of sizes from 3 to 5 mm at $£ 2.93$. The fourth tool set, (model 37305) comprises two cross point screwdrivers, three hexagonal key wrenches ( $1 \cdot 5,2$ and 2.5 mm A.F.) and tommy bar at $£ 3.93$.


Light Soldering Developments tool sets.

Addresses of nearest 'stockists can be obtained from Light Soldering Developments Ltd., (Dept. EE), 97-99 Gloucester Road, Croydon, Surrey.

## CONSTRUCTIONAL PROJECTS

## EE Radio Control System

Our star project this month is part one of the EE Radio Control System series and obviously will call for some special components. These will be described fully in the various articles.

Apart from the special electromechanical items, the majority of components should be generally available. The special components are usually stocked by local radio control shops, but any readers experiencing difficulties can order them from S.L.M. (Model) Engineers Ltd., Dept EE, Chiltern Road, Prestbury, Chelten. ham, Glos, GL52 5JQ.

## 3-Function Generator

The only item likely to cause concern in the 3 -Function Generator is the integrated circuit IC1.

We have found that the 8038 is only available from Maplin Electronic Supplies or through R. S. Components dealers.

## MW/LW Radio Tuner

For the MW \& LW Radio Tuner, the slow motion (Jackson ' O' gang type) tuning capacitor is listed in the Maplin, Watford and Home Radio catalogues. However, the specified coils seem to be rare and only stocked by Home Radio Components.

## Baby Alarm

The 741 integrated circuit used in the prototype model of the Baby Alarm was a TO-5 can type with preformed leads. The 8 -pin d.i.l. plastic package is more common and readily available and can directly replace the can type.

Quite a number of readers will already possess a high impedance microphone so therefore the mic. insert could be omitted and SK1 chosen to suit your mic. plug.

The use of a rotary switch for S1 is optional and any double-pole toggle switch will suffice.

## Opto Alarm

The first in our Uniboards series is a simple Opto Alarm.

There are numerous solid state buzzers on the market at the moment and it is worth shopping around for this item as prices seem to vary quite considerably.

The thyristor type MCR"102 would appear to be only available from Maplin but the 2N5060, 2N5061 and 2N5062 types are suitable replacements.

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## with Kelth Cadbury



## USEFUL SUCKER

ONE of my main sources of high quality components for stock is the "Goody Bag". Whenever I visit my local electronics shop, I rummage in his "junk" bins and usually select a bag or two of assorted "goodies".

Until recently the various p.c.b.'s that I had collected from these bags of components had been gathering dust. Most of the components on the boards had leads too short to cut, and removing them with a soldering iron proved to be one hell of a laborious task, resorted to only in emergency, when a particular component has been needed that was not available from another source.

A recent acquisition has resulted in all the boards being stripped of 75 per cent of their components, and at a very fast rate. I now have a stock of several hundred close-tolerance resistors, items which have previously been bought only as required.

The acquisition that made it all so easy was a device called a "Soldersucker'. A sort of suction device with a Teflon nozzle, it can be primed and discharged with one hand easily, while the other hand is used to apply the soldering iron to the soldered component. The Soldersucker draws away molten solder with fantastic force that has to be seen to be believed, and after repeating the operation at each of the joints, the component can be lifted out, sometimes without the need to heat the "de-soldered" joints again.

So simple and so quick, I just didn't realise how easy its use makes the removal of components. I would not have considered spending over a fiver on the tool, but as I have now had the chance to prove its worth at, relatively speaking, no cost (it was amongst a large "job lot" I was fortunate enough to obtain for a few quid recently) I have no hesitation in recommending its worth.

It would soon cover its cost. I have recovered, in good order, something like eighty pounds' worth of transsistors, 1 per cent resistors, integrated circuits and capacitors, with the aid of the Soldersucker!

DREAM of an electrontc house, where everything is controlled from central position. Heating, lighting, ventlation, entertalnment, security, cooking wathing and so on,

To sit In a Captain Kirk-type of armchalr and to be in complete control of one's immediate environment seems to me to be quite possible given today's state of the Art. And glven the time and the money to make it all!

A robot to take the dog for a walk; three VTR's always recording all TV output, recalled by ultrasonic instruction at a moment's notice for replay on one of the many colour televisions around the house; similar audio recorders for five or six radio programme transmissions; automatic tending of the garden. What bliss, but for how long, before the whole caboodle becomes an absolute bore? You would get no exercise ever, and you would possibly die of a heart
attack brought about by the effort of rising from your control chair to go to bed.

Nevertheless, those readers who dream of more electronlckary will realise the necensity of a patchboard, to alter varlous parameters that may need adjustment-how long grandma is allowed in the bath before tho water automatically drains away; grilling times for the T-bone steaks; securing the fridge and freezer when hungry teenagers go prowling.

Even more modest projects will benefit from a patchboard-it would be an additional item of equip. ment that could prove very useful to the enthusiast's audio set-up, espec.ally where creative tape-recording is undertaken.

The patchboard described here is adequate for all projects the writer has worked on to date, and can be made at a fraction of the cost of a "bought" item.
 lot of money to spendl For example, the Maplin catalogue price quoted for a $30 \times 30$ hole patchboard is £ $88 \cdot 38 \mathrm{p}$. It seems to me that my-very-cheap alternative would suffice in nine out of ten applications.

Chassis mounting phono sockets are avallable on Paxolln boards containing numbers of sockets from one to elght, from Maplin, and work out at under 5 p per socket in most cases. For example, to make an alternative to Maplin's $10 \times 10$ hole board costs under a fiver, using twenty of the five-socket boards, compared with £19.55p.

I used single strands of copper wire, about $1 \frac{1}{2} \mathrm{~mm}$ thlck, from a length of electricians' heavy-duty cable, which was soldered as shown in the lllustration. Careful drilling and mounting of the boards is needed to make the finlshed job
look neat-but then care is needed with all electronic work anyway!

And that's not all-the plugs are much less expensive also. Ordinary phono plugs cost under 10p, and can either be shorted out, or small reslstors or capacitors can be connected across the terminals, Inslde the cover. Use plastic plugs (which are the cheapest) and devise your own colour code so that you can tell at a glance whether the connectlons are shorted or Jolned through a component. The wireable component plugs listed by Maplin for their $10 \times 10$ board cost 59p each, compared with 9p for my alternativel

Yer pays yer money and takes yer cholce-for me, Mr Hobson dictates, prompted by the bank manager, tax collector, starving children and shoeless wife.


## Wireless Telegraphy Act

The legality of remote and radio control understandably confuses many people. Here are the facts in a nutshell. The Wireless Telegraphy Act prohibits the use of any unauthorised radio station.
This wording covers both trans. mitters and receivers. So it is not only illegal to transmit any radio frequencies (such as CB radio) without authorisation, it is also illegal to receive them.
It follows that it is also illegal to use a radar speed trap detector in a car. These devices pick up police radar speed check signals and convert them into an audible alarm.

Under the Wireless Telegraphy Act it is also illegal to use a radio controlled model boat, car or aeroplane. But whereas no authorisation and licences are available to transmit pirate radio programmes or receive police radar signals, licences are available for the transmission of non-speech radio remote control signals to models and toys.

The penalty for any illegal trans. mission or reception, whether Citizens Band chat, radar trap avoidance, pirate tradio pop music transmission or radio remote control of a toy, is the same; a fine of up to $£ 400$ and/or 3 months in jail. It is, of course, highly unlikely that anyone using a remote control toy would be fined as much as someone transmitting a pirate radio programme, but the penalty is available to a court.

## Direct Link

Fortunately, because the Wireless Telegraphy Act covers only radio frequencies, it does not cover the use of ultrasonic, or infra red, or visible light, or laser light, links for remote control or other communication, even of speech and music. Thus it is perfectly legal to use links of this type without a licence. The snag is
that such links are far more directlonal than radio línks.

In Japan it Is now possible to buy a gramophone turntable that contains a built-in high quallty stereo radio transmitter which operates on a v.h.f. f.m. band. The gramophone signal can thus be picked up by a v.h.f. f.m. receiver anywhere in the house. So the user can install a turntable in one room and an amplifler and $h l f i$ system in the other without any cable IInks. This would be Illegal in the UK.

Ultrasonic or Infra red llnks need something close to Ilne-of-sight relationship, so cannot offer a comparable facility. Also infra red Ilnks can be disturbed or "broken" by direct sunlight, as the sun emits considerable infra red radiation.

I recall eyewltness tales of an Im. pressive demonstration several years ago which was set up to show off the prowess of a remote controlled fire fighting device. The robot-llke gadget was designed to sense the Infra red radiation produced by a flre, turn, drive towards It and then loose off the contents of a fire extingulsher.
The demonstration took place out of doors and a can of petrol was duly ignited. It was Summer, but a dull day. Then, just as the petrol burst Into a ball of flames, the sun broke through the clouds. The robot's sensor plcked up the sun's infra red radiation and latched onto its direction. The gadget stopped dead in its tracks, tilted back and loosed the contents of Its fire extinguisher into the sky.

## Take-away Car Radlo

In-car-entertalnment or ICE Is now big business. It's easy to pay around $£ 300$ for a combined radlo and cassette player; and that's excludling loud. speakers, and extras like booster amplifiers, graphic equallsers and exotic aerials.
Understandably many motorists are reluctant to install such expensive
equipment because it's akin to leaving several hundred pounds laying in the dashboard pocket ready for a thief to grab. Even worse, the thief will probably smash the door, break a window or slit your sunshine roof to get access.

- Burglar alarms are one answer, but by no means 100 per cent. Another answer is that offered by car radio firm Voxson.

The Voxson Tanga range of radios, now being fitted as standard to small Fiat cars, is the very opposite of secure. The radio is a plug-in module that the driver removes every time the car is left unattended.

The really clever part of the scheme is that they have made the removable module small enough to fit into a pouch that hangs on a key ring along with the car keys. A socket is secured to the car dashboard and as this socket contaıns only a single chip audio amplifler It Isn't worth stealing. The tiny plug-In module contains all the r.f. and i.f. circuitry, a tuning control and a volume control. There's a separate colour-coded module for longwave, medium wave and v.h.f. reception.

Provided you remember to pult out the module when you park there's nothing left to encourage a thief.

## War on CB

I learned recently how CB helped us win the war in Africa. Of course it wasn't called CB then, but the wavelength, 27 MHz was the same.
Before World War II such fre. quencies seemed unmanageably high. But spurred on by the Impetus of war the USA, Japan and Germany all made military equipment to work on this band.
One of the characteristics of " 27 meg', and indeed one of the reasons why no one wants it for CB In the UK, Is that it can skip across Continents. A signal beams up into the sky boun. cing off the upper atmosphere and down to earth again thousands of miles away.
In 1942 an amateur radio enthusiast in the USA heard German conversations on his experimental 27 meg receiver. He brought in a German speaking friend who reckoned the conversation sounded like military chat between tank commanders.
The American army moved in and dlscovered that the signals were skipping across the world from Rommel's tanks in North Africa. They could only be plcked up within a radius of a few miles and night after pight the army in the USA monitored the signals from Africa and sent them back to Field Mar shall Montgomery in Africa. Thus, although Montgomery was out of range of Rommel's low power 27 MHz transmitters, he soon knew everything the "Desert Fox" was saying to his troops.


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# RADIO WORLD <br> <br> , 

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## By Pat Hawker, gзva

## Amateur News Service

For over 24 years, a specialised "broadcast"' news service entirely independent of the BBC and IBA has quietly but efficiently existed in the United Kingdom: the RSGB's weekly "GB2RS' bulletins transmitted every Sunday morning from amateur radio stations in different parts of the country. The bulletins provide news and information of interest to all radio amateurs and short-wave listeners.

An important extension to this service has just been introduced: the bulletin now, for the first time, goes out at 1100 hours local time on $7 \cdot 0475 \mathrm{MHz}$ using conventional amplitude modulation and can thus be heard by listeners with run-of-the-mill "all-band" radio receivers.

Previously all GB2RS transmissions have been on 3.5 or 144 MHz , often using single-sideband or narrow-band frequency modulation, frequencies and modes seldom available to listeners not equipped with communications receivers designed specifically for radio amateurs.

The 7 MHz transmissions will usually come from the station of Gordon Adams, G3LEQ at Knutsford, Cheshire and reception in the UK will depend on the "short skip" conditions to be expected at this stage of the sunspot cycle.

Apart from 7 MHz the new schedules include seven transmissions at different times from different sites on 3650 kHz ( 3640 or 3660 kHz in Scotland) using ssb or a.m.; eight transmissions on ssb on $144 \cdot 250 \mathrm{MHz}$; and 19 transmissions on 145.525 MHz nbfm , together providing coverage in most parts of the UK.

The service was launched in September 1955 by Frank Hicks-Arnold, G6MB on behalf of the Radio Society of Great Britain. Since then one of the London news-readers, Arthur Milne, G2MI of Bromley, Kent has read the bulletin on more than 1000 Sundays; he can usually be heard making the first transmission on $3 \cdot 65 \mathrm{MHz}$ each Sunday at 0930 local time.

A condition imposed by the Home Office is that the weekly scripts, prepared at RSGB headquarters, have to be vetted by them in advance. Bulletins provide details of national and international happenings and events affecting amateurs, contest results, propagation conditions, news of amateur expeditions ("dxpeditions'), OSCAR satellite orbital predictions and the like.

There is also a weekly bulletin for radioteleprinting (rtty) enthusiasts transmitted under the call-sign GB2ATG in the $3 \cdot 5$ and 144 MHz bands-of course on radio teleprinters

In these days when there is much in. terest in the concepts of local and com. munity radio broadcasting, GB2RS provides an interesting example of an alternative concept: that of reaching nationally a relatively small segment of the population. By using their own communications transmitters the radio amateurs have shown a way of doing this at low cost.

## Radiation Non-hazards

Events at the Kensington fire station, where in August radiation meters appeared to detect harmful levels of ionizing radiation but where it was shown by staff of the National Radiological Protection Board apparently to have been caused by harmless non-ionizing radiation from the short-wave transmitters of the nearby Israeli Embassy, have underlined once again how difficult it is for the lay public (and even the experts) to judge iust what levels and types of radiation are potentiaally harmful.

Most scientists and engineers accept that the present officially recommended levels for non-ionizing radiation from microwave and other radio transmissions, even though set empirically many years ago, have proved remarkably satisfactory, though there still remain doubts in some minds as to possible biological effects at levels too low to cause appreciable local heating.

Contrariwise there are some grounds for thinking that low levels of h.f. radiation may even have a beneficial, preventive effect in regard to certain diseases.

## Microwave Bombardment

Part of the confusion in the public mind was brought about by the much publicised "'bombardment" by microwaves of the US Embassy in Moscow some years ago. Many people rushed to the conclusion that this was all a deliberate attempt to affect the health of the American diplomats.

Less well known is that it has become clear since then that the real reason was a Russian attempt to prevent interception of their microwave telecommunications links by receivers in the Embassy, a practice they were themselves doing in the USA. There is considerable evidence that their embassies and consulates contain microwave aerials and receivers which can intercept telephone traffic to and from Government buildings, using computers programmed to select automatically conversations likely to be of interest.

Many embassies, of course, have h.f. radio transmitters that enable the diplomats to communicate directly with their own countries. My daily walk to work through Belgravia takes me past several large and very prominent "log-periodic" h.f. beam arrays, while even a casual look at many of the other diplomatic buildings in the area reveal more modest transmitting aerials. And some countries still favour "disguised" aerials, hidden in flag poles, etc either in deference to environmental considerations or as a relic from the days when diplomatic radio links were virtually a form of under-cover "pirate" operation.

Today it is all highly "legal" under Article 27 of the Vienna Convention on Diplomatic Relations which gives to missions the right of free communication
in code or cipher, although still insisting that missions "may install and use a wireless transmitter only with the consent of the receiving state". Occasionally problems arise from the transmitters causing interference to television reception in the area, a matter which has to be handled with diplomacy.

## Why So Slow?

Among the reasons why so many hobbyists would welcome a CB system are the difficulties, the delays and the expense of obtaining an amateur radio licence. It takes too long and costs too much for a youngster to acquire a Class A or a Class B amateur licence. It is not just a question of the technical standards but also the administrative delays. Now that the Radio Amateurs Examination is based on "multiple choice" questions, capable of being marked very rapidly, why is it usually September before candidates learn whether they have passed an examination held in May? And why do candidates have to apply to take the examination so long beforehand?

With a sufficiently large pool of multiple choice questions it should surely be possible to arrange that applicants could take the exam at any time, virtually on a walk-in basis, just as those living near a Post Office coast station or Marine Radio Surveyor's Office can take the Morse test at any time of the year. Time seems so very important to a youngster itching to get on the air.

I was fortunate enough, as a schoolboy, to take out my licence before there was such a thing as a technical examination but considerable technical interest in radio communication

In these days of factory "appliances" there is a lot to be said for checking that applicants do know something about the technology-but nothing at all to be said for putting such long delays into the system.

Further evidence of the value and importance of encouraging amateur radio emerged in the aftermath of the floods in west India and in the path of Hurricane David in Dominica where for a period the only link with the outside world was via an amateur station operating from batteries.

"Just a moment while I get my calculator".


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# By Harry T. Kitchen 

## Marking-Out

Last month I advocated the creating of a drawing, however elementary, of the required marking-out; I also explained the reason for doing this in reverse. Let us now look at marking out, cutting, and bending a fictitious front panel. In real life, of course, you will substitute your own requirements.

Let us agree on a front panel measuring 10 in by 6 in , and let us work in imperial since so many of us do so in our private lives, whatever measurements we may use at work. Let us also decide that the panel will be secured to the cabinet by means of flanges $\frac{1}{2}$ in wide, bent inwards, and at right angles to the panel. Immediately this gives us the overall size of 11 in by 7 in . We cut this from a larger sheet of aluminium, or obtain it cut to size.

All four sides will, naturally, be absolutely square. We must mark out our datum lines, commencing with the two centre lines. Set the combination square to $5 \frac{1}{2}$ in and scribe a small line; likewise at $3 \frac{1}{2} \mathrm{in}$. Using the square, now extend these lines until they intersect, bang in the centre of the panel,dividing the sheet into four exactly equal portions.

The position of every hole, top to bottom, side to side, is, in good engineering practice, referred back to these centre lines, so their exact positioning is critical. So too is every bending line. Errors are thus confined to one reference line.
Now if we happily start at one end and carry on, line to line to line, errors can accumulate, possibly disastrously. Say every line is out by 25 thou., in itself a wide or a narrow limit depending on applied criteria, then six holes, or lines, later on you will be ou't of position by $0.025 \mathrm{in} \times 6$ or 0.150 in . That hole or line being out of position could completely ruin the panel.

## Fixing Flanges

The fixing flanges require a somewhat different approach. If you mark the panel to precisely 10 in by 6 in it will not fit. Why? Well, you haven't allowed for the thickness of the metal. For a precision panel you must subtract the thickness of the panel from the bending dimensions.

In round figures let us say the panel is 25 thou. thick. So you set your combination square to 5 in and 3 in from the centre lines, and then as well as you are able to, you subtract 25 thou. each time, top and bottom, and both
sides. Then scribe the bending lines. With decent luck you will achieve a panel that is a perfect fit. When bentl
Now we can set about the holes required. Round holes are easy; at the intersection of appropriate horizontal and vertical datum lines use a centre punch and lightly "pop" the precise point. Then use engineers' dividers to draw the circle required. Square or rectangular holes also use the horizontal and vertical datum lines. Locate the centre of the hole then, halving the width and length scribe its limits above and below, and to either side of the datum lines.

Let me reiterate that these lines will have been scribed on the reverse side of the panel so that the outer side is unblemished when the panel is completed. Got it wrong? So have I before, and I dare say, will again.

## Cutting Out Holes

Having a panel marked out, we can commence cutting out the holes. There are various tools on the market designed to facilitate this chore. Let us however confine ourselves to easily and cheaply obtained hand tools. Of inestimable value is the Abrafile, available in various diameters. I have had mine for many years, and they range from $\frac{3}{4}$ in diameter to some that will fit a fretsaw; just the job for cutting holes in metal panels.

For round holes, drill a starting hole just inside the circumference of the required hole somewhat larger than the Abrafile, or other round file you propose using. Insert your file and away you go, all around the hole, just inside the scribed cricle. Enlarge the hole to the required size, and remove all rough edges, by use of a smooth half round or round file. Smaller holes are simply enlarged in size by judicious use of a round file.

Square or rectangular holes are tackled in a similar manner. Again a starting hole is drilled, this time in one corner. Again you set off with your trusty round file, filing away just inside the scribed lines. Finally you square off the corners and straighten up the sides by use of a smooth Hand file or Flat file.

Alternatively, you can, particularly with large holes, drill several holes in a straight line, inside and parallel to each side of the hole. Then you use a padsaw with a length of hacksaw blade in it to cut out the hole. The four sets of holes you drill must, of course,
all join up so that the hacksaw blade can be inserted. Finish off as before.

## Bending

The scribed bending line must be accurately aligned with the angle iron, and just visible. This degree of visibility is important as it aids repeatability. The angle iron pieces are bolted together and clamped in the vice securely.
Use a piece of hard wood, place it in intimate contact with the aluminium sheet and the angle iron, and bend the sheet in the same direction as the scribed lines until it is flush with the angle iron; hopefully this will be square.

If necessary, tap the hard wood with a mallet, from end to end, and back, slowly and carefully. When the sheet lies on the angle iron, place the hard wood upon it and tap it down firmly to ensure a good tight bend. There should be no signs of damage on the sheet, or ripples; the hard wood used as an inter-face between mallet and sheet is a great aid here as it absorbs local blows.

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| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
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| ${ }_{211}^{111}$ | 0.5 1.0 | ${ }_{0}^{0.25}$ | 2. 2.4 | 0. |
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| 㖪 | 8 | 3 | - 5 | 0 |
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| 116 |  | ${ }^{6}$ |  |  |
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