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| $100 \mu F$ | $25 V$ | $10 p$ |
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| $250 \mu F$ | $50 V$ | $17 p$ |
| $500 \mu F$ | $25 V$ | $18 p$ |
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| $1,000 \mu \mathrm{~F}$ | 25 V |
| :--- | :--- |
| $1.000 \mu \mathrm{~F}$ | 50 V |
| $2.000 \mu \mathrm{~F}$ | 25 V |
| $2.000 \mu \mathrm{~F}$ | 50 V |
| $2.500 \mu \mathrm{~F}$ | 25 V |
| $2.500 \mu \mathrm{~F}$ | 50 V |
| $3,000 \mu \mathrm{~F}$ | 25 V |
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## everyday electronics

 THEORY.
## FOR ALL SEASONS

Our cover this month has quite an outdoor touch. Of course, you don't have to be an apiarist to sense that things are beginning to buzz in the outside world. Spring is now well advanced and thoughts are likely to be turning towards all kinds of pastimes and occupations for the coming summer months.

It is an appropriate time to point out that doityourself electronics has no closed season. Outdoor activities like gardening, touring, camping, sporting events, and so on, present many unique opportunities for putting electronics to effective use. So we advise, take stock now, anticipate your needs and start building to remedy any deficiencies in this respect.

## GOOD COMPANION

The Constructors Companion given free with every copy of this month's Everyday Electronics is small and compact. It has been designed for your pocket, so that wherever you go you can have essential facts constantly at hand. Compiled with the beginner particularly in mind, this booklet will prove a valuable aide-memoire for the more experienced constructor as well.

Those still feeling their feet will be glad of the technical back-up they can instantly call upon
when confronted with a choice of allegedly alternative or equivalent parts when shopping personally for components.

## READY ACCESS

Our regular readers will already appreciate the amount of important and useful information they are accumulating, as the months go by. True, not everyone will have an immediate need for every project described. But a word of advice: do not discard back numbers. You never know when circumstances may arise that create a definite need which some previously described project would satisfy exactly.

This leads us on to another common problem:. how to store numerous copies of a magazine so that ready access may be made at any time to one particular article. The only really satisfactory solution is to keep copies of the magazine in the binder specially designed to hold 12 issues of Everyday Electronics. and which is now available.


Our June issue will be published on Friday, May 19
EDITOR F. E. BENNETT - M. KENWARD - B. W. TERRELL B.Sc.

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## EASY TO CONSTRUCT SIMPLY EXPLAINED

VOL. I NO. 7 ..... MAY 1972
CONSTRUCTIONAL PROJECTS
AUDIO TONE GENERATOR Suitable for making electronic music by F. C. Judd ..... 358
BEE COUNTER An electronic eye for the hive by G. A. Cozens ..... 376
METAL LOCATOR Detects and locates metal objects underground by D. Bollen ..... 382
GENERAL FEATURES
EDITORIAL ..... 356
SHOP TALK New products and this month's constructionals by Mike Kenword ..... 362
MAKING ELECTRONIC SOUNDS AND MUSIC Using the Audio Tone Generator by F. C. Judd ..... 363
please take note ..... 367
THEY MADE THEIR MARK No. I-Introduction By J. E. Gregory ..... 368
TEACH-IN Part 7-Semiconductors: Transistors by Mike Hughes ..... 369
RUMINATIONS by Sensor ..... 374
TEACH-IN HALF TERM TEST ANSWERS See how well you did ..... 389
READERS LETTERS Your news and views ..... 390, 393

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Fig. 1. Complete circuit diagram of the Audio Tone Generator.
ratio will be 1 to 1 or the mark and space will be of similar duration (Fig. 2). If we now change one of the controlling values-in this case VR1both the frequency and mark to space ratio will be altered.

If we increase the value of VR1 the frequency will decrease as C2 will take longer to charge, and the mark to space ratio will alter for the same reason (see Fig. 3). Thus frequency control is achieved by VR2 and the total frequency range is approximately 50 to $2,000 \mathrm{~Hz}$.

The waveform has a mark to space ratio of 1 to 1 at approximately $1,500 \mathrm{~Hz}$ at all lower frequencies the mark to space ratio increases becoming about 1 to 20 at the lowest frequency (Fig. 3).

The output from the multivibrator is taken from the emitter of TR2, through R6 to the base of TR3. Transistor TR3 is switched hard on and off by the output from TR2 and thus ensures a completely square output at its collector. The output level from TR3 is continuously variable from 0 to approximately 7 volts by VR2.

## SAWTOOTH OUTPUT

The square wave output from TR3 can also be switched via S2 through an integrating network, C5 and R8, to provide an approximately sawtooth waveform (Fig. 4) of about 1 volt peak-to-peak maximum output, instead of the square-wave.

One of the major differences between a square wave and a sawtooth wave is the harmonic content and hence the tonal quality, when either are made audible via an amplifier and loudspeaker. The square wave contains only odd harmonics, in addition to its fundamental,
whereas a sawtooth wave consists of both odd and even harmonics plus the fundamental.

Audibly the square wave has a sound rather like that produced by a clarinet particularly in the region of middle C ( 261 Hz approx.). The sawtooth wave has a sound rather more like a


Fig. 2. A square wave with a 1 to 1 mark to space ratio.


Fig. 3. A square wave with a 1 to 20 mark to space ratio.


Fig. 4. A sawtooth waveform.

## Audio Tone Generator



Fig. 7. Wiring of the final unit. The tinted area is the component mounting board as shown in Fig. 6.



Fig. 6. Top and underside views of the component board. The transistor connections between the two diagrams are viewed from the underside.
flute. Both waveforms are used extensively in electronic organ voicing and for electronic music.

## CONSTRUCTION

The prototype unit was housed in a box made from universal chassis parts. The pieces used assemble into a box measuring 5 by 4 by 3 inches. The sides and top and bottom can be assembled leaving one plate for the front panel and one for the rear. The plate used for the front panel is drilled as shown in Fig. 5 and is used to mount all the components.

If the layout and assembly of the generator is as shown there is just room in the case for a PP9 9 volt battery. Even if you spread the layout a little there should still be room for a slightly smaller 9 volt battery. The circuit board is 0.15 inch matrix plain perforated veroboard and is mounted on a 2 inch length of $3_{8}$ by $3_{8}$ inch aluminium angle.

## Components....

## Resistors

| R1 | $2 \cdot 2 \mathrm{k} \Omega$ |
| :--- | :--- |
| R2 | $47 \mathrm{k} \Omega$ |
| R3 | $4.7 \mathrm{k} \Omega$ |
| R4 | $3.2 \mathrm{k} \Omega$ |
| R5 | $100 \Omega$ |
| R6 | $4.7 \mathrm{k} \Omega$ |
| R7 | $4.7 \mathrm{k} \Omega 2$ |
| R8 | $15 \mathrm{k} \Omega$ |
| All $\frac{1}{6} \mathrm{~W}$ | $\pm 10 \%$ carbon |

## SEE SHOD TAHK TALK

## Capacitors

| C1 | $50 \mu \mathrm{~F}$ elect. 12 V |
| :--- | :--- |
| C 2 | $0.05 \mu \mathrm{~F}$ |
| C3 | $0.01 \mu \mathrm{~F}$ |
| C 4 | $50 \mu \mathrm{~F}$ elect. 12 V |
| C 5 | $0.1, \mu \mathrm{~F}$ |

Transistors
TR1 NKT 274 germanium pnp
TR2 NKT 274.germanium pno
TR3 NKT 274 germanium pno
Potentiometers
VR1 $1 \mathrm{M} \Omega \log$ carbon
VR2 $2 k \Omega$ lin carbon
Switches
S1 S.P.S.T. slide
S2 S.P.D.T. slide

## Miscellaneous

SK1 Phono socket
B1 PP9 9V battery
Control knobs (2 off) Eagle type F10, case $5 \times 4 \times 3$ in made from universal chassis panels or a similar size case, battery connector, aluminium angle $2 \times \frac{\frac{3}{8}}{} \times \frac{3}{3} \mathrm{in}$. Veroboard $5 \times 4 \times 0 \cdot 15$ in matrix plain perforated, earth tag, connecting wire, 48A fixings.

Commence wiring of the component board by inserting all components except the transistors, and the wire link on the top of the board as shown in Fig. 6. Turn the board over and connect up the two supply lines along the two sides of the board using 18 or 22 s.w.g. tinned copper wire. Next connect up the remaining components using the component leads where possible and connect the flying leads.

Finally insert the transistors checking carefully the lead connections with the underside view shown in Fig. 6, and solder them to the other components using a heat shunt on each lead as it is soldered. After checking the circuit board mount the board on the aluminium angle bracket and mount this on the front panel together with the remaining components.

Wire up all the components to the circuit board as shown in Fig. 7 and check the completed unit carefully before connecting the battery and switching on.

Continued on page 386


This month we have one item which many readers will prabably wish to construct but which is not given in the form of a constructional project. It is the simple passive mixer that is described and drawn up in the Making Electronic Sounds and Music feature.

Since this is really a bonus that will be useful to those follow. ing the article we have not given full constructional details or a components list. All the component values are given on the circuit diagram and the wiring diagram shows how they are put together. The three sockets can be any type suitable for use with your particular tape recorderthe types we have shown are phono sockets.

The complete unit can be mounted in any small case. No battery or power supply is necessary. We would like to emphasise that this is a simple passive mixer and will not be able to cope with all inputs.
A more advanced type of mixer may form the subject of a future article. However this simple mixer should be suitable for use with the Audio Tone Generator that is also described in this issue.

## Audio Tone Generator

There should be very few buying problems for the Audio Tone Generator. As described above the sockets could be changed to any suitable type if your equipment does not use phono sockets or if you already
have other types. Once again the case for this project can be any type that is available in a suitable size.

## Bee Counter

We find it difficult to comment on the availability of cedar wood -not after-shave-but apparently this wood must be used or the bees will not accept it!

As far as the remaining components for the Bee Counter go make sure that the resistors you buy are of adequate wattage. The lamp and holder should be of the miniature type so that they can be accommodated in the wooden base panel. Since the current drawn by this circuit is fairly large the section in the article concerning the battery should be noted.

There are a number of Post Office type counters available so make sure you get the right one -4.2 ohms coil resistance is the important thing.

## Metal Locator

The Metal Locator is a project which we are sure will create great interest but please remember that this is a simple onetransistor design and cannot be expected to out-perform a $£ 30$ unit. The use of Perspex or Paxolin is recommended for the locator head as these materials are not affected by damp or water.

All remaining components for the locator should be readily available. The use of a subminiature switch is recommended since only a small hole then needs to be cut in the plastic beaker. Any $50 \mu \mathrm{~A}$ moving coil meter could be used in the locator provided it will fit the beaker lid. The one specified is probably the cheapest.

Finally do not forget the operating licence and don't say we did not tell you!

## New Products

Two products from one goahead firm have been introduced this month. Both in the audio field, possibly the most outstanding is the Unisound 505 as Radio and T.V. Components call their do-it-yourself $£ 25$ stereo system. This competitively priced unit comes as a complete kit and only needs two screwdrivers to put together. All the electronics are in module form and are supplied
with wiring looms that only need connecting up using a screwdriver supplied with the kit.

The large EMI speakers are housed in attractive cabinets again put together with only a screwdriver. It is said that anyone who can wire up a mains plug can put the system together in one evening. The system utilises modified Mullard Unilex modules, has an output of 3.7 watts continuous sine wave r.m.s. per channel; and frequency response of 40 Hz to 20 kHz at the 3 dB down points. It would be very difficult to buy the individual components -including Garrard 2025TC deck, cartridge, plinth and cover and build a unit to match this one for $£ 25$, excluding the two speakers and cabinets.


The second unit from RT-VC is a $£ 7$ push button car radio kit, slightly more difficult to construct but any reader who has some experience of soldering should be able to build a working unit.

The kit is of good quality and uses the same push button tuning unit as radios costing three or four times the price. These features ensures good sensitivity and the pre-aligned i.f. (intermediate frequency) module and tuner avoid complicated alignment.

The kit is suitable for 12 V positive or negative earth operation and readers may like to note that an after sales serviceto repair any item not functioning correctly-is operated by RT-VC for all their kits; cost about $£ 2$ depending on the fault.



## Simple experiments with a tape recorder

THE term "electronic music" almost defies explanation because it is not the music that is electronic but the equipment and methods of ereating it.

Its origin goes back many years, in fact to the invention of the thermionic valve and even as early as 1921 a "concert" of electronic music was performed in Paris by an Italian, Luigi Russolo, who used what was then called electrical sound generating and reproducing equipment.

Electronic music was difficult to perform directly from sound generators, etc., because composition required arranging the sounds in a given order and even changing the order, and sometimes the sounds, at a later time.

## MODERN METHODS

Magnetic tape recording finally provided the ideal medium for composition. The sounds required could be recorded and rearranged afterwards by simply cutting out the pieces of tape containing them and splicing these together again in the order required. This technique paved the way for composers who, with both electronics and magnetic tape at their disposal, could produce new kinds of music with tonal qualities never before possible.

More recently of course the music synthesizer has taken over the task of tone generation, etc., and electronic music composers can now programme a synthesizer, couple it to a tape recorder and produce "instant" electronic music.

Nevertheless there is much that can be accomplished by the amateur with an ordinary domestic tape recorder, an audio tone generator (like the one described on page 358) and some splicing tape. The techniques are simple and you can get a good deal of fun out of experimental electronic music and "science fiction" sounds.

Your efforts need not be wasted either because you can enter them for the experimental music and sounds section of the annual British Tape Recording Contest (details later).

## EQUIPMENT

An ordinary spool to spool tape recorder is the main requirement and if you have a stereo recorder with provision for recording independently on either track or you can get together with a friend and use two tape recorders, so much the better. A tape recorder with track-totrack or duoplay facility is also advantageous especially if it permits echo effects.

It is not possible to lay down procedures for specific makes and types of tape recorder but you will find that most of the techniques described can be applied.

Note that cassette or cartridge tape recorders are of limited use for creative recording of this nature which requires fairly extensive tape cutting and splicing.

Most modern spool to spool tape recorders are designed for stereo operation employing half or quarter track on standard quarter inch wide tape. If the tape recorder has a track-totrack recording facility it will have separate recording and replay heads, thus allowing a recording on one track to be copied on to another together with other signals.

Some stereo recorders may only have a common record/replay head which will not normally allow track-to-track copying but may have a facility for making separate recordings on each of two tracks. Information concerning such facilities should be given in the tape recorder instruction book. If in doubt, you should contact your dealer or the manufacturer for such information.

## AUDIO TONE GENERATOR

An audio tone generator is not absolutely essential but is most advantageous. The simple Audio Tone Generator described on page 358 is quite suitable as it covers a wide enough frequency range and will deliver a square-wave or a nearly sawtooth-wave output signal, thus providing two basic sounds.

Sounds picked up by a microphone can also be used because these can be reshaped by tape cutting and splicing and by certain recording techniques. Magnetic tape will be required of course and for initial experimental work low priced brands will suffice.

Some splicing tape and blank leader tape will also be required. Do not use ordinary plastic sticky tape, such as Sellotape, for splicing as

Fig. 1. (a) Original waveform of the recorded sound (b) The sound recorded and shown in (a) played in reverse.

(a)

(b)

Fig. 2. Waveform of a sound which starts instantly and slowly dies away.
this may damage the tape and will not give a long lasting joint. Small kits of coloured leader and proper'splicing tape are readily available. A small tape splicer is also a very useful, though not essential, tool.

## FIRST EXERCISES

It is important to know the extent to which your tape recorder can be used. If it has two or three speeds, as most of them do, record some musical sounds, whistling will do, at all three speeds and then play them back at one speed only, say the highest.

The sounds recorded at the lower speed(s) will be raised in pitch, by one or two octaves, depending on the speed. If the replay speed is double that of the recording speed the pitch is raised one octave and the sounds will occur faster but if the replay speed is half the recording speed, the pitch will be reduced by one octave and the sounds will occur slower. This is one of the most simple but most used techniques.

## REVERSE REPLAY

Now, if your recorder is a stereo machine try turning the tape over (reverse the spools) and see if you can obtain replay on another track in reverse, i.e., the sounds will be going backwards. This technique is also commonly used for electronic music because it alters the nature of the sound completely by placing what was the beginning of the sound, i.e., its attack, at the end as illustrated in Fig. 1 in which (a) is the sound as recorded and (b) as played in reverse.

If you cannot play sounds in reverse try this exercise; connect a tone generator, or if this is not available record whistles through a micro-

Photograph showing the use of a tape splicer to join up a number of sounds.



Recording various sounds, using the microphone, to form a composition.
phone. Start with the recording level control at the maximum, set the tape running to record the sound but then almost simultaneously slowly turn the record level control to zero.

On replay you will have a sound that starts instantly and then slowly dies away as in Fig. 2. With a little practice you will be able to get various dying away or decay times depending on the speed at which the recording level control is turned off. Now try the reverse procedure; gradually increase the sound whilst recording and then quickly stop it.

## TAPE CUTTING EXERCISES

Now try some tape cutting; first use the highest tape speed and record a few sounds of different pitch, i.e., from a tone generator, or whistles via the microphone, each one lasting two or three seconds.

Locate the beginning of each sound on the tape by carefully feeding the tape across the head and then cut the tape about two inches in front of the sound. Run off the remainder until you reach the beginning of the next sound; cut the tape here and splice to the end of the piece containing the first sound. Cut and join pieces of the remainder of the sounds.

On replay you will have a series of short sounds each rapidly following the other. Now try a similar exercise but this time insert pieces of blank leader tape between each sound.

## mUSIQUE CONCRÊTE

Finally a variation of the two previous exercises. Record a few sounds each at a different tape speed. These should preferably be musical sounds, such as whistles or tones, or sounds produced by tapping a wine glass for example. Cut one or two pieces of each from the tape and assemble them at random with pieces of blank leader between groups. The pieces may be long or short.

Try replaying the assembled tape at different speeds and note the effect. You are well on the way to a form of composition known as "musique concrête" which is the creation of abstract forms of music out of real sounds. The same technique can, however, be used for abstract forms of electronic music in which the main sound source is an audio tone generator.

## USING A TONE GENERATOR

The exercises outlined above demonstrate how almost any recorded sound can be altered by tape cutting and by recording and replay at different tape speeds. Electronic music does not normally include natural sounds recorded via a microphone and therefore the sound sources are electronic, i.e., from tone generators of one kind or another. The recording and tape cutting techniques, however, remain the same.

If you have a full range audio signal generator then tones can be recorded at the pitch required. The simple generator described on page 358 has a frequency range of approximately 50 to 2000 Hz .

If frequencies outside the range of the generator are required it is simply a case of recording and replaying at different tape speeds for example; if a frequency of around 4000 Hz is required, record the highest pitch of the generator (approximately 2000 Hz ) at a tape speed of $3^{3}{ }_{4} \mathrm{in} / \mathrm{sec}$ (inches per second) and replay at $71_{2} \mathrm{in} / \mathrm{sec}$.

If a very low pulsing sound is required at say 20 to 25 Hz record a square-wave signal from the generator at its lowest pitch and then replay the recording at half the speed. Some experiment in this direction will soon reveal the tonal and pitch ranges that can be obtained simply by recording and replaying at various tape speeds.

Once this has been done, further experiment with the audio tone generator can be carried out in order to discover the type of sounds that can be produced. Start by recording a continuous note and while recording this vary the frequency and output controls on the generator, try this for both the square and sawtooth outputs (note that the output in the sawtooth position is much lower than in the square wave position).
Try cutting and reversing the sounds recorded to obtain various effects. You can also try making recordings at a distorted level by turning up the record level control, this will distort the original sound and produce yet another effect. Try switching from one output waveform to the other whilst recording-you can vary frequency and output at the same time-and also try switching the generator on and off while recording, again you can vary the output and frequency whilst turning on and off.
Edit the sounds produced by cutting and splicing and experiment fully with all possible


Fig. 3. Circuit diagram of a passive mixer that can be used for making electronic music.
effects. Once you have done this and feel fully conversant with the various effects that the generator is able to produce you can start to add one effect to another.

## SIGNAL MIXING

Recording from track-to-track or using two separate tape recorders may necessitate mixing signals that are to be recorded and re-recorded i.e., signals from a recording already made to be mixed with signals from another source such as the tone generator.

Some recorders have built-in mixing facilities whilst others may permit a form of mixing by using the track-to-track recording facility or by superimposing one sound on another previously recorded. Again the tape recorder instruction book will provide information of this nature.

However, it is possible to build a very simple mixing circuit as shown in Fig. 3; Fig. 4 shows the construction. This is known as a passive mixing network, but will alow two signal sources to be mixed at different levels and coupled to a common input on a tape recorder (Fig. 5).

## TAPE LOOPS

Another interesting technique widely used for electronic music is the tape loop. This is the use of a small endless loop of tape containing recordings which are played continuously to produce repeating rhythm patterns.

Record a few natural sounds, or low pitched tones from an audio generator, of quite short duration, one immediately after the other. Cut a piece of the tape containing the sounds, about 18 inches long, and splice the ends together so as to form a complete loop. Place the loop in the recorder so that it runs past the tape heads when the machine is set to replay. You can hold the loop under tension by one of the methods shown in the photographs. Try running the loop at different speeds and, if possible, reverse the direction.

Record some percussion sounds, e.g., sounds produced by knocking together empty boxes, etc. Cut out pieces and make up a loop consisting of the various sounds and blank leader tape.


Fig. 4. Constructional details of the circuit shown in Fig. 3. Shop Talk refers to this figure.

For the first attempt use only two or three sounds and two or three pieces of leader.

You can make up an almost endless variety of fascinating rhythm patterns by this method and if you use two tape recorders the rhythm loop can be copied from one to the other whilst other sounds are added.

## MULTIPLE RECORDING

If you have a tape recorder with a track-totrack recording facility the scope is much wider as sounds may be recorded on one track and then re-recorded on to another track whilst adding more sounds. If your tape recorder can produce the echo effect this too can be used in various ways to produce those echoing science fiction sounds. Try allowing the echo to build up into a crashing roar and see if you can play it in reverse.

Now that you have discovered the variety of sounds and rhythms available using the facilities you have it is up to you to put these together to form an interesting "musical" passage. It may take some time before you achieve the required effect.

By combining even a few of the techniques outlined the number of permutations possible are fantastic. Instructions on composition cannot be given because no rules exist. Your ideas must come solely from imagination and experiment.

Fig. 5. Using the passive mixer to combine two signals for recording purposes.



The three photographs on the left illustrate various methods of using a tape loop. The top photograph shows a reversed loop held under tension by passing it around a pencil ; this is only suitable for short periods.
The centre photograph shows a reversed loop held under tension by a small spool hanging over a table edge; this is only suitable for fairly large loops.
The lower photograph shows a system that can be used for any size loops by routing the tape around suitable objects-batteries are shown. This photograph also shows a cardboard tape holder used to keep recorded sections of tape in the order required.

## COMPETITION

Finally, why not try an entry for the "technical experiment class" of the annual British Tape Recording Contest. It is open to anyone and the closing date for the 1972 contest is not until June 30. The Technical Experiment class allows for tapes of up to 4 minutes duration and includes; sound composition, electronic music, musique concrête, multi-track music and experimental sound recordings. The prizes are worthwhile and you can get an entry form free by writing to The Secretary, British Amateur Tape Recording Contest, 33 Fairlawnes, Maldon Road, Wallington, Surrey, and enclosing a stamped addressed envelope. You may also be interested to know that the special "Tape of the Year" award for 1971 was for an experimental class entry.

Every tape entered is carefully assessed by the expert judges and their comments are passed to the contestant concerned when the tape is returned. Thus you will know how to make an even better tape next time.

## PAEAE NOTE

The approximate cost of components given in the Simple Calculator article last month was incorrectly shown as $£ 1 \cdot 20$. This should have been $£ 2 \cdot 20$.

The probe flying lead in the Signal Injector article (March issue) should be soldered to Y3 not $Y 2$ as stated in the text.

The Normatest 2,000 multi-range test meter mentioned in Shop Talk last month is avallable from: Croydon Preclsion Instrument Company, Hampton Road, Croydón, CR9 2RU.

# TH MADE THEIR MARK NO1 Introduction By J. E. Gregory 

ELECTRONICS is an internationally uniform world of symbols. Look at any advertisement or study the simplest circuit diagram in Everyday Electronics and you will be confronted with strange symbols of every shape. Magical signs used to signify basic units of physical quantity; Table 1 lists some of them.

Although electronics is regarded as a modern science and hobby many of these units are named after pioneers, scattered throughout the world, whose accumulated research spans hundreds of years.

This series sets out to explain the symbol, and perhaps more important something of the man who gave his name to it. But let's begin our potted history of electronics at the beginning

## THE GREEKS HAD A WORD FOR IT

Take the word electronics itself, for that we must go back in time to ancient Greece. To the ladies of Greece passing time by decorating their spinning wheels with amber, found on shores, in the far north. They observed that the amber when contacting the threads would draw the threads to itself as they separated from the wool, and then push them away in a frictional force. The

Greek word for amber was elecktron, from the verb elkein to attract. Although this phenomenon was observed and noted by several of the great Greek philosophers we have to jump two thousand years to the early 1600 's and to the reign of Good Queen Bess, who was persuaded by her physician William Gilbert, to attend a demonstration of a frictional electric machine based upon the power of amber to attract. This power he called electricity.
It was soon realised that the crackling and sparking of Gilbert's electric machinte were the same phenomena on a minute scale, as thunder and lightning, but how to prove it?

## THE KITE FLYER

One of the first to try was the fifteenth child of an English immigrant; born in Boston Massachusetts in the year 1706, this was the well known American statesman and philosopher Benjamin Franklin (see illustration above).

His historic but dangerous
experiment trying to capture electricity from the sky accurred during a thunderstorm in the summer of 1752 , when accompanied by his small son, he flew a kite with an iron door key. During the storm, he saw that sparks sprang from the key to his wrist, what he didn't realise of course was that if the lightning had actually struck the kite he would have been killed.

The study of natural phenomena had to take second place to his other activities, but he came to the conclusion that thunderstorms were simply the levelling of opposed electrical potentials, between one cloud and another or between a cloud and earth.
It was Franklin who introduced the positive and negative signs for electric charges, realising there are two kinds which neutralise each other.
Next month we move from America to 18th Century Italy and a scientist, Alessandro Volta, after whom the Volt, the measurement of electrical potential is named.

Photograph: Sclence Museum. London.
Table I FUNDAMENTAL UNITS

| unit symbol | name of unit | Physical quantity |
| :---: | :---: | :--- |
| $\mathbf{A}$ | Ampere | Electric Current |
| $\mathbf{V}$ | Volt | Electric Potential |
| F | Farad | Electric Capacitance |
| $\Omega$ | Ohm | Electric Resistance |
| $\mathbf{W}$ | Watt | Power |
| Hz | Hertz | Frequency |
| $\mathbf{H}$ | Henry | Inductance, |

These basic units are often inconveniently large or small and the units are prefixed with the following symbols:

| $p$ | pico | $\div 1,000,000$ million |
| :--- | :--- | :--- |
| $n$ | nano | $\vdots I, 000$ million |
| $\mu$ | micro | $\vdots I$ million |
| $m$ | milli | $\vdots i, 000$ |
| $k$ | kilo | $\times 1,000$ |
| $M$ | mega | $\times I$ million |
| $G$ | giga | $\times 1,000$ million |

Hence $5 k V=5,000$ Volts ; or $5 \mathrm{mV}=0.005$ Volt

ELECTRONIC CIRCUITS-
IN THEORY and PRACTICE


By Mike Hughes M.A.

THis year sees the twentieth birthday of the component most responsible for bringing electronics within the scope of do-it-yourself enthusiasts; it has greatly simplified design and construction and has also brought about terrific reductions in costs. It is the "transistor".

As a replacement for the valve, it allows us to use low voltages and removes the arduous task of having to assemble valve bases and massive transformers on tank like chassis. Connections to a transistor are few and the basic way it operates in a circuit is quite easy to understand.

PNP-NPN
The transistor is a member of the semiconductor family and is basically a sandwich of different types of either silicon or germanium. The "filling" of the sandwich can either be p- or $n$-type material; we can clad a $p$ - type filling with $n$ type material giving what we call an $n p n$ transistor. Alternatively a pnp device is made by filling a $p$ - type material with an $n$-type.

One encounters both types in practice but nowadays $n p n$ devices made from silicon predominate, the reason being that they are easier to make and hence cheaper!

Fig. 1(a) shows a diagramatic cross-section of both types of transistor, $p n p$ and $n p n$. One end is heavily doped and is called the "emitter"; the other end is lightly doped and called the "collector".

The filling material is very thin in practice (usually one or two microns; 1 micron is a
millionth of a metre) and is called the "base". In its simplest form you can think of an $n p n$ device as two diodes connected together by their anodes (back-to-back), and facing each other in a pnp device, Fig. 1(b).

(a)

(b)

(c)

Fig. 1. (a) Schematic diagram of the internal make-up (b) equivalent representation and (c) circuit symbol for (top) npn transistor and (bottom) pnp transistor.

## BASE CONNECTIONS

All the transistors you will come across have connections brought out from the emitter, base and collector. A very common silicon npn device is the BCl 08 and we shall be referring to this frequently in this series.

Fig. 2 shows what it looks like. If you have one handy see if you can identify which lead is which.


Fig. 2. Top and underside views of the BC108 npn transistor showing lead connections. The metal case is internally connected to the collector.


The emitter is the one closest to the spigot on the side of the can, the collector is diametrically opposite, and the base is between the two but set off to one side. This is a metal can transistor and the can is electrically "live"-in actual fact it is connected to the collector as well as the lead out wire.

Different types of transistor may have different shaped cans and some are in plastic encapsulations. Always make sure you know which lead is which before you start using a transistor.

Most constructional projects in Everyday Electronics give you lead designations for the transistors specified, but if you want to experiment with alternative types make sure you know the correct base lead connections.

## SIMPLE TEST

Use the $\mathrm{BC108} \mathrm{npn}$ transistor to identify the effect of the two diodes connected back-to-back. First of all make an ohmmeter on the Demo Deck. Use a 4.5 V battery (not 9 V ) in series with a $2 \cdot 2$ kilohm resistor and VR2 ( 5 kilohm). Complete the circuit and set VR2 to give zero ohms at full scale deflection and then connect the leads of your ohmmeter between the base and emitter connections of the transistor-to do this it is best to solder the transistor on to three adjacent pins of the Demo Deck and use crocodile clips on the leads from the meter.

If you connect the meter so that the lead coming directly from the negative terminal of the battery goes to the emitter, the meter needle will move to almost full scale showing there is little resistance in the transistor. Now reverse the leads so that the base is more negative than the emitter-you should see that no current
flows (indicated by meter needle not moving). Thus the base-emitter junction is a diode and follows the same rule that we saw last month.

Now leave the lead on the base and transfer the one from the emitter to the collector-again no current flows but reverse the leads and current flows between the base and collector.

If you connect the leads between the collector and the emitter no current should flow whichever way you have them because in both connections, the current would have to pass through a reverse biased diode.

This simple experiment can be used as a rough and ready test to check if a transistor is likely to be in working order, and provided you remember the rule "make p stand for positive for current to flow" you can use it to identify $n p n$ and $p n p$ transistors.

## REVERSE VOLTAGE LIMITS

Like all diodes, the junctions of a transistor have reverse voltage limits. These are usually specified with abbreviations. For the BC108 the reverse emitter-base voltage ( $V_{\text {ebo }}$ ) is 5 V -i.e. you must never make the base more than 5 volts negative with respect to the emitter (this is why we had to use 4.5 V for our ohmmeter instead of the 9 V we have been used to). Likewise the reverse base/collector voltage ( $V_{\text {cbo }}$ ) is 30 V . You might expect the reverse voltage between the emitter and the collector to be equal to the highest of the other two but this is not the caseit is lower-for the $\mathrm{BCl} 108 V_{\text {reo }}$ is 20 V .

The " 0 " in the suffixes of the reverse voltage characteristics indicates that the third terminal is "open circuit" i.e not connected.

## HOW THE TRANSISTOR WORKS

Let's see what a transistor actually does by using the circuit of Fig. 3(a). Now that we are using the transistor in a real circuit it is important to note the polarity of the supply volt-age-for an $n p n$ transistor the collector must always be kept more positive than the emitter (the converse applies to pnp devices). We are going to make the transistor work like a tap and control the amount of current flowing through R1. You can see this happening if you follow the details through on the Demo Deck.


VR1 is a 300 ohm potentiometer working as a potential divider giving us a variable supply at its wiper.

Wire up the circuit of Fig. 3(a) on the Demo Deck as shown in Fig. 3(b), but do not connect R2 to the base of the transistor just yet.

Resistor R3 and the 1 mA meter makes a 10 V range voltmeter in the usual way. Connect the negative lead to the emitter of the transistor. All voltages we measure will be relative to that of the emitter.

First measure the power rail at point X-it should, of course, be +9 V ; now measure the potential at the collector of the transistor (point Y ) it should be $+8 \cdot 2 \mathrm{~V}$. This is what is expected because no current can flow through the back-toback diodes of the transistor, but the meter will draw some! If you had a high sensitivity meter (say 20 kilohm per volt) this current would be negligible and you would see +9 V at both points, X and Y .

Now set VR1 so that the potential on its wiper is zero (with respect to the emitter) and connect R2 to the base of the transistor. VR1 potential is measured by attaching the crocodile clip from the meter to point Z. Again measure the potential at the collector-it should not have changed.

We shall now see what happens if we increase the potential at the wiper of VR1. Do this in 0.5 V increments (use crocodile clip at point Z ) and for each setting measure the collector potential. You should see that once the potential of the wiper exceeds 600 mV , the potential at the collector falls, and continues to fall towards zero as the controlling voltage is increased. Once the collector potential reaches almost zero no more
control can be effected. We say that the transistor is now fully conducting between collector and emitter. This state is called "saturation."

Record your results and plot a graph of collector voltage versus voltage at the wiper. A graph should be obtained similar to that of Fig. 4.


Fig. 4. The graph obtained by plotting the recorded results of experiment using circuit of Fig. 3(a), i.e. voltage at point $Y$ versus voltage at point $Z$.

Control of the collector/emitter current is brought about by passing a current through the forward biased base/emitter junction. The more current we pass into the base in this way, the more current we can control between the collector and the emitter. The controlling current is called "base current," ( $I_{\mathrm{b}}$ ) and the controlled current "collector current," ( $I_{\mathrm{c}}$ ).

Base current is set by the potential difference between the wiper of VR1 and the emitter of the transistor, acting through the resistance R2

Fig. 3(a) (left). The circuit diagram used for investigating some of the properties of a BC108 transistor.

Flg. 3(b) right. The circuit of Fig. 3(a) wired up on the Demo Deck.

and any internal resistance between base and emitter. The latter is small and can be neglected at this stage. We must, remember, however, that the base must be made at least 600 mV positive with respect to the emitter before any current can flow (this is the usual forward voltage drop for any silicon junction).

We can thus calculate the current flowing into the base by measuring the potential at the wiper of VR1, subtracting the base emitter forward voltage drop $(600 \mathrm{mV})$ and dividing by the value of R2.

## GAIN

If you do this for your experiment you will find that the base current ranges from 0 to $0 \cdot 084 \mathrm{~mA}$. The range of collector current we are controlling was from 0 to 9 mA . It can be seen that the transistor enables us to use a very small current to control a larger one. We call this effect "current amplification." The factor that governs the ratio between $I_{\mathrm{b}}$ and $I_{c}$ is called "gain" and although it increases with $I_{c}$ it is pretty well constant for any given transistor. It can, however, vary widely between different types of transistor and even between devices having the same type number! Provided you take a combination of base and collector currents within the controllable region (this is called "linear region") you can calculate the gain of the transistor you are using.

It would be best to increase the potential at VR1 until the collector potential is approximately 4 V . This reduces the shunting effect of our voltmeter.

Use the precise values of voltage measured to calculate the current through R2 and R1 then use the ratio of these values to calculate the gain.
gain $=$ collector current $\div$ base current $=I_{0} \div I_{b}$
For the BC108 transistor it should be approximately 200 , but as we have said, will vary from device to device.
Example To calculate the gain from your plotted curve (similar to the one of Fig. 4) select a convenient point. on the linear region such as point $P$ of Fig. 4.

The base current, $I_{\mathrm{b}}$ is given by the voltage difference between the base and emitter divided by the base resistor.

$$
\text { i.e. } \frac{3-0 \cdot 6}{100.000}=0.024 \mathrm{~mA}
$$

Now the voltage drop across the collector resistor Rl is $(9-4) \mathrm{V}=5 \mathrm{~V}$. Therefore, collector current $I_{\text {s }}$ is $(5 \div 1000)=5 \mathrm{~mA}$.

Substituting these values for $I_{\mathrm{c}}$ and $I_{\mathrm{b}}$ in equation (1) gives the gain $=(5 \div 0: 024)=208$.

There are various ways of describing current gain for a transistor so we shall define that measured above a little more precisely-it was the d.c. current gain. This is sometimes abbre-
viated to the designations $\beta$ (beta) or $h_{\text {Fe. }}$ The latter is rather a strange type of designation but is one of a range of what are called " $h$ " para-meters-we need not worry ourselves about these in this series except for the term $h_{\text {PE }}$ which is usually used in manufacturer's data sheets. Do not confuse $h_{\mathrm{Fe}}$ with $h_{\mathrm{te}}$, the latter is called the small signal current gain and we shall not be dealing with this until later.

The gain equation above can be rewritten: $I_{\mathrm{c}}=h_{\mathrm{FR}} \times I_{\mathrm{b}}$

Remember that the experiment we have just done has been using a silicon npn device. We could have used one made from germanium having $n p n$ structure and obtained a similar effect-except that the base/emitter forward voltage drop would have been only about 200 mV and $h_{\mathrm{PE}}$, in general, would have been lower.

We could also have used a silicon or germanium pnp device but would have had to reverse the battery connections so that the collector was negative with respect to the emitter. The same rules would have applied and we could have still calculated a value for $h_{\text {PE }}$.

If you are a little confused by the difference between $n p n$ and $p n p$ devices do not worry too much as this stage-most of the early experiments in Teach-In will use $n p n$ devices and when you have got used to these you will find it quite straightforward to switch over to $p n p$ devices when necessary. The most important thing to remember is the polarity of battery voltage when using one type or the other. An aid to remembering what the polarity ought to be is to bear in mind the direction of conventional current flow;


Fig. 5. Circuits showing major current flow directions for (a) npn and (b) pnp transistor. $I_{b}$-base current, $I_{\mathrm{c}}$-collector current, $I_{\mathrm{e}}$-emitter current.
the arrow on the emitter of the symbol points in the direction of current flow, i.e. it points away from positive and towards negative. See Fig. 5.

Whether using $n p n$ or $p n p$ devices an aid to remembering how to turn collector/emitter current "on", is to make the potential at the end of the resistor connected to the base tend towards the same polarity voltage as applied to the collector; the more you move towards this voltage, the more $I_{\mathrm{b}}$ increases, and $I_{0}$ will increase in direct proportion.

When the potential feeding the base rises towards the supply voltage the voltage at the collector falls towards the emitter voltage. This is called "inversion."
In Fig. 3 R1 is called the "collector, load." The limit of $I_{\text {e }}$ control is set by the value of this resistor; if it has a high value then it does not matter how much base current you apply, you cannot control more collector current than that given by the collector supply voltage divided by the value of collector load. On the other hand, if the load is too low you might find yourself trying to force more collector current than the construction of the transistor can handle. Thus one of the specifications of a transistor is the maximum collector current it can handle without "blowing". This is called $I_{\text {emax }}$ and for the BC108 is 100 mA .
A final parameter we must deal with is the power rating of a transistor. As current is passing through it a certain amount of heat is dissipated. We already know that too much heat can spoil the properties of a semiconductor so it must be limited. The limit is set by the maximum power dissipation parameter, $P_{\text {emax }}$. It is easy to calculate what the power dissipation is likely to be; it is the dissipation you would get if you replaced the transistor in the circuit with a resistor having the same ohmic value as the collector load.
Table 1 gives you some typical values of parameters for some common transistors of varying types, powers and polarities.

Table 1: THE MORE IMPORTANT CHARACTERISTICS OF SOME COMMON TRANSISTORS

|  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 300 mW |  |  |  | 100 |  |
| 92 | np | 200 m | 18 V | 18 | 5 V | 100 | 150 |
|  | $n p n$ | 800 m | 60 | 60 | 6 V | IA | 70 |
| $\times 13$ | pnp | 300 mW | -20V | $-15 \mathrm{~V}$ | -5V | 100 m | 120 |
| 3702 | pnp | 360 mW | -40V | -25V | -5V | 200 m | 60 |
| ACl26 | pnp | 500 mW | -32V | -32V | -10V | 10 | 00 |
|  | $p \mathrm{p}$ | 125 mW | -16V | -16V | -3V | 125 mA | 50 |
|  | pnp | 12W | - 16 | 16 | -10V | 3.5A |  |
|  | pnp | 30 | - |  |  | 10A |  |

Fig. 6(a) (below). The circuit diagram of the "Electronic Candle" which illustrates positive feedback.
Fig. 6(b) (right). The circuit of Fig. 6(a) wired up on the Demo Deck. Ensure that PCC1 is close to LP1.


## ELECTRONIC CANDLE EXPERIMENT

We shall now make a simple working circuit using the circuit diagram of Fig. 6(a). This is wired up on the Demo Deck as shown in Fig. 6(b). Connect the ORP12 (light dependent resistor) very close to the LP1 on the Demo Deck as shown below. Set VR2 to zero ohms. The potential at the base of TR1 will be zero, therefore no current will flow between collector and emitter. Now, in a reasonably lit room, increase the value of VR2. At a certain point the potential at the base will reach 0.6 V (set by the potential dividing effect of PCC1 and VR2) and the transistor will start to conduct (the bulb will glow dimly).

Continue to increase the resistance of VR2; the current flowing through PCC1 will now pass into the base/emitter circuit of the transistor in preference to the higher resistance path through VR2. This base current will cause TR1 to pass more collector current until the bulb is fully illuminated.

When you reach this point (the minimum value of VR2 that will give full illumination) try casting a shadow over PCC1, the lamp will go dim and ultimately go out altogether as $I_{b}$ reduces due to the resistance of PCC1. We did a similar sort of thing in Teach-In Part 4.
The difference is that we now have a circuit that is much more sensitive to small changes in light level which is brought about by the transistor amplifying the current from the photo resistive cell.
If you place the cell very close to the bulb in a dimly lit room you can set the value of VR2 so that the ambient lighting does not turn the transistor on, but the light from the bulb will.


Break the light path between the bulb and the cell and the bulb goes out and stays out. Now use a match or lighter to provide a stimulus of light. Bring it close to the bulb/cell assembly and the bulb lights up; you can now remove the match and the bulb will stay on because its own light output is holding the transistor on. This is called "positive feedback" and in this circuit will provide an amusing party trick-especially if assembled to look like a candle.

A bit of practice at "snuffing" the candle with the fingers (actually you are breaking the light path between the bulb and the cell) will make the effect even more astounding.

Photograph of the Demo Deck set up for the Electronic Candle Experiment showing the lamp being "lit" by the light emitted from the lighter.


## TEACH-IN PART 6-ERRATA

Fig. 4(b) last month shows a lead connected wrongly. The lead from the junction of R3 and the negative meter terminal should go to the negative end of VR1 (not the wiper as shown) i.e. the one connected to the battery negative.


Next month: Multivibrators. The components needed for next month in addition to those already acquired are: resistors 22 kilohm ( 2 off), 100 ohm (1 off); capacitors $0.1 \mu \mathrm{~F}$ polyester ( 2 off), $500 \mu \mathrm{~F}$ elect. 12 V (1 off); transistors BCI 08 (I off); diodes OA91 (I off).

## Ruminations By Sensor

## Not so Clever

The coal miners' strike has shown how dependent we are, in this age of high technology, on the efforts of men who work in damp, dirty and often dangerous conditions.

I find it difficult to comprehend that on one hand the semiconductor industry owes its existence to the ability to obtain and to process materials with an impurity content of less than ten parts in a thousand million, and to operate with tolerances down to one millionth of a metre, while on the other hand men have still to dig fossil trees out of the earth (albeit with mechanical assistance) so that these fossilised remains can be burnt to boil water in order to raise steam
and to generate electricity! Without coal and electricity there would be no semiconductor industry; truly our idol has feet of clay!

## Let There be Light

Have you heard about the old lady who telephoned the C.E.G.B. to complain that, during the power cut, the buses were passing her house with all their lights on? She also said that she could manage to get along quite well without the electricity, except for the little light in the hall, and could they please leave that one switched on.

Many people must have been irritated, in the early days of the strike, to see street lights blazing all day and switched off at night, due to their electric clock switch mechanisms getting umpteen hours behind. To the electronics man the answer to this problem is so simple-a light operated switch, either using discrete components or in integrated form.

A recently introduced inte-
grated circuit provides the necessary photo cell, level sensor and time delay all on one tiny chip of silicon and complete with lens. It could operate a relay or, better still, work into a switching transistor controlling the street lamp directly.

Some years ago, I was shown around a large generating station, where, tucked away in a dusty corner there was a cast iron box about the size of a domestic cooker. This apparatus was installed at the station about twenty five years ago and its purpose was to switch on all the electric street lamps in the town.

When switched on it produced a ripple which was superimposed on the mains. Sections of street lighting were grouped together under the control of master switches, spread throughout the town, which were operated by switching on the ripple equipment. The system had been in use but for some reason, unknown to my guide, had been discontinued. It would have been a blessing during February 1972.


## Use this 'electronic eye' to monitor the busyness of your bee-hives

##  C

MODERN research calls for accurate measurement and comparisons, and with this in mind this device was designed to help the beekeeper assess the performance of his beehives more definitely, and to compare the different strains of bees under the same working conditions and so help to breed a strain which will produce the most honey under all the difficulties encountered in our changing climate without the rather nasty habit of the English bee, of attacking the bee-keeper as soon as he appears anywhere near the hive.


The Bee Counter is an instrument which records the number of bees entering the hive, and used in conjunction with other devices such as a wind speed indicator, a wind direction indicator, an air temperature thermometer, a maximum/minimum thermometer, a rain gauge and a sunshine recorder, then some degree of assessment can be made, and some basis established for the bee-breeder to work upon his main goal-lots of honey from a reasonably good tempered, busier bee.

The Bee Counter makes use of the fact that bees are highly organised in their habits, and utilises the bees sense of sight and smell. These bee "characteristics" are used in the design of the cabinet housing all the circuitry which is described later in full detail.

## THE CIRCUIT

The complete circuit diagram of the counter is shown in Fig. 1 and is basically an amplifier which works as follows.

The lamp LPI, which is always "alight" when


Fig. 1. The complete circuit circuit diagram of the Bee Counter.
the unit is switched on, illuminates the light dependent resistor, PCC1, and causes its resistance to be at a low value, about 100 ohms.

The l.d.r. and R2 form a potential divider circuit and under "illuminated conditions" of the l.d.r., a positive voltage with respect to the emitter, is applied to the base of TR1 causing it to be in a conducting state.

With TR1 conducting, a negative voltage is applied to TR2 base with respect to the emitter and consequently TR2 is "off" (not conducting).

When the light path between LP1 and PCC1 is broken, the resistance of PCCl increases considerably (to about 100 kilohm for complete "blackout"). This causes the potential at TRI base to go negative and turns it "off". This state of TR1 causes the voltage applied to the base of TR2 to go more positive and causes it to switch "on" i.e. conduct-current flows through TR'2.

When current flows through the emitter leg of TR2 containing the relay coil in the counter, the relay is energised.

When the light to PCCl is restored, TR2 switches "off" and the counter is de-energised and springs back to its off position, and in doing so mechanically adds "one" to the counter readout.

The arrangement of LP1 and PCC1 in the case is so devised that the bee, on entering the hive, breaks the light path between these devices and its entry is thus recorded.
The $13-20$ ohm 3 watt resistor, R5, in the collector circuit of the power transistor, TR2, is to prevent damage to the counter or the transistor if the entrance passage to the hive should become blocked, as once happened in the prototype. when a drone got stuck in the narrow part.

A heavy duty battery is required to operate the Bee Counter since current drain is substantial -250 mA when TR2 is "off" and 400 mA when TR2 is "on" at 12 V . A car battery is therefore recommended to supply the power. The cost of this battery is not included in approximate cost.

The voltage is fairly critical as it must be sufficient to operate the counter, but not high

## Components....

## Resistors

R1 20-56S2 3 watt
R2 68ks2
R3 $1 \cdot 2 \mathrm{k} / 2$
R4 $1.2 \mathrm{k} \Omega$
R5 $13-20 \Omega 3$ watt
All $\frac{1}{2}$ watt carbon SEE stated
$10 \%$ unless otherwise
ransistors
TR1 OC72 (or similar) germanium pnp TR2 OC26 germanium pnp

Light Dependent Resistor PCCI ORP12

## Micellaneous

LP1 12-14V 0.75 W bulb and holder
PL1, PL2 Wander plugs, 1 red 1 black (2 off) SK1, SK2 Sockets to suit plugs PL1, PL2 B1 12V battery-heavy duty rechargeable type (Not accounted for in cost box.)
Counter: Post Office type 14C $4 \cdot 2 \Omega 4$ figure readout. Cedar wood, Perspex and adhesive, Paxolin, wood screws, 4 B.A. nut and bolt, wood glue.
enough to cause overheating of TR2 or the counter coil in the event of the passage being blocked for long.

If the apparatus is disconnected every night the battery will last at least a week on one charge.



Fig. 2. The layout of the components on both sides of the Paxolin board. Veropins are used for attachment.

Variations in performance can be dealt with in several ways. The lamp should be bright enough to turn off the amplifier, but not any brighter than necessary. This is best adjusted by altering the series resistor R1, which may be increased to as high as 56 ohms.

Also, the size of the light hole can be varied, or a part of the l.d.r. painted over so that it has less area exposed, until the instrument is sufficiently sensitive, but positive in its action.

## THE COUNTER

The electromagnetic counter used is a Post Office type. It has a four digit readout and can thus count up to 9,999 . The maximum count rate is ten per second.

## COMPONENT WIRING

Most of the components of Fig. 1 are mounted on a piece of Paxolin size $4^{1}{ }_{2} \times 1^{1}{ }_{4}$ inches with a cut-out as shown along one side to accommodate the light dependent resistor, PCC1.

Both sides of the board containing the components are shown in Fig. 2.


Veropins are used for mounting the components in position and small holes should be drilled where indicated to accommodate these pins.

Three more small holes of the same size should be drilled to take the leads of TR1 as shown.

Drill the component board fixing hole and the four holes for transistor TR2; (see reverse side of component board Fig. 2); ${ }^{1}{ }_{8} \mathrm{in}$. diameter holes will do for all five holes.

Begin assembly by pushing in all the Veropins and then attach TR2 to the board using two small nuts and bolts.

The connection to the collector of TR2 is via its casing, so a solder tag should be attached to one of the securing bolts to enable this connection.

Attach and solder all the components, link wires and flying leads as detailed in Fig. 2 making sure a heat shunt is used when soldering in TR1, which incidently should be the last component connected.

The l.d.r. should be attached to the board via 6 in . long flexible leads.

The flying leads to the counter should be about 4in. long.

The two wander sockets used for battery connection to the counter, are attached to the end of the case which is made from a piece of Paxolin, dimensions are given in Fig. 3.

The connection wires from the wander sockets to the component board should be about 4 in. long.

Connection to the battery is made via two wander plugs and a length of twin flex

Fig. 3. Dimensions of the end made from Paxolin to accommodate the wander sockets for battery connection, and counter readout.



A photograph of the prototype with top and tunnel lid (which holds PCC1 in position) removed. The photograph clearly shows the entrance and exit tunnels (labelled IN and OUT respectively). The take-off platform, made from Perspex, is located just beneath the exit cut-out, and is glued in position with Perspex adhesive.


## EXIT AND ENTRANCE GEOMETRY

As said before, this device and its design utilises the bees' senses of smell and sight: From inside the hive, the exit from the hive appears as a bright opening to the outside world and so the exit path through the instrument must be a tunnel with transparent sides and top to allow this condition to be fulfilled.

In the instrument this tunnel slopes upwards so that when the bee emerges, it finds itself on a platform of Perspex, about ${ }^{3}$ in. wide, situated above the hive base, and flies away.

When it returns, it will land on the hive base (landing/alighting board) and walk towards the hive.

The entrance to the hive is now through the Bee Counter which is a tunnel painted matt black; when the bee walks along the front of the instrument and reaches this tunnel it will enter.

On entering, the tunnel becomes narrower and at the same time slopes upwards until it is just wide enough for a single bee to pass.

There is a lamp under the narrow part, with a hole in the floor of the tunnel, made up to the level of the floor with Perspex cement so that light can shine up through it.

The light dependent resistor is situated in the
roof of the tunnel and as the bee walks between this and the lamp, the light beam is cut and the circuit activated.

Positions of the "electronics" within the case.



## CONSTRUCTION OF CASE

Cedar wood should be used to construct the case as this material will be readily acceptable to the bees.

Cedar wood will also withstand the weather without the need for painting but it is well to remember that if the counter is to be used in exposed outdoor conditions, weather protection becomes an important consideration, whereas in laboratory conditions it is not so.

The best compromise for an outdoor installation is a shelter which will keep off the rain.

First of all make all the wooden parts of the case as detailed in Fig. 4.

Now solder the two thin flexible covered wires to the bulb holder tags and screw in the bulb. These wires are led out through the top of the base and the bulb assembly is glued in position.

It is not likely that the bulb will need replacement because it is "under run" and there is a 20 ohm resistor (R1) in series with the bulb which reduces the light and heat dissipated in the bulb.

When the glue has set, fill up the light hole with Perspex cement so that it comes flush with the passage floor.

Glue down the two sides of the tunnel so that the width of the narrowest region is ${ }_{4} \mathrm{in}$. Paint the tunnel top, bottom and sides a matt black.

The light dependent resistor should be a push fit into a hole in the tunnel roof.

Glue and screw the front and back to the base and glue the exit ramp in position. Drop the tunnel roof into position indicated. The other parts of the case are made from Perspex and their dimensions are given in Fig. 4.

With these made we can proceed with the assembly.

## ASSEMBLY

Begin by screwing the Perspex side and top windows in position as indicated. Glue the Perspex platform to the front and place the Perspex exit guide in position.

Now solder the two wires from the bulb holder to the component board as detailed in

Fig. 2, push the l.d.r. in position and then attach the board to the back of the case by means of a 4 B.A. nut and bolt. This bolt should be countersunk into the back so the back is flush with the front of the hive. If there is a gap here, the bees will try to go in or out through the smallest crevice.

Attach the wander sockets to the Paxolin side and solder to the appropriate flying leads from the component board. Next screw the Paxolin side to the case.

When the flying leads to the counter have been connected, fit the counter into its locating holes, (one end in the Paxolin and the other in the bracket on top of the Perspex exit guide) and secure with nuts. The counter digits should be visible through the slot in the Paxolin side.

Screw the top on and the unit is complete.

## CAPACITY AND POSITION OF CASE

The single entry counter (as this is) is only suitable for a three or four frame hive, since with a full scale hive the returning bees would sometimes overload the tunnel capacity.

The maximum a single entry counter can handle is about 60 per minute.

For a full scale hive a three entry counter is necessary. This means the entry tunnel is divided into three passages, each with its own light beam arrangement, amplifier and counter.

Whereas the single entry model is only $6{ }_{4}{ }_{4} \mathrm{in}$. wide, which is about right for most observation hives, it is better to make the three entry model $16{ }^{1}{ }_{2}$ in. wide so that it takes up the whole width of a Standard National hive.

When the counter is put in front of the hive the hive should be moved back by a distance equal to the depth of the Bee Counter, in this case $3^{1}{ }_{4} \mathrm{in}$. so that the point of entry is exactly as it was without the counter.

When this is done the bees will soon get used to the new conditions and will be using the exit and entry passages without any confusion.


## A simple, easy to construct selfcontained locator giving a meter indication of buried metal.

By. D. Bollen

## Approximate Itcomponents... £ 4:00 plus case

The metal locator described in this article was designed for simplicity and ease of operation. A single transistor circuit is used to give a clear meter indication of the presence of buried metal without the need for headphones or a nearby portable radio as used by some locators. Under typical operating conditions the instrument will detect a 2 p coin at a depth of about 1 inch.

## CIRCUIT OPERATION

The complete circuit of the metal locator is shown in Fig. 1. Transistor TR1 acts as a common base oscillator with positive feedback between collector and emitter controlled by trimmer C4. Search coil inductor L1 is tuned by C 2 to give an oscillation frequency of 100 kHz .

When the circuit is functioning, Ll will induce eddy currents in nearby metal and this transfer of energy causes an increase of TR1 emitter current. Although small, the accompanying change of d.c. voltage across R3 can be detected by a sensitive null (or zero registering) voltmeter.

In Fig. 1 the d.c. null voltmeter consists of R3, R4, R5, R6 and. ME1. Capacitor C5 is included

## 10 (2) AT (as) (2)

to remove unwanted a.c. from the voltmeter input, and diodes D1 and D2 protect the meter movement against overload.

At a certain setting of C4, the d.c. voltage at TR1 emitter will equal the voltage at the junction of R5 and R6 so that no current flows through ME1; this can be taken as the normal operating point for the circuit. If metal is brought close to L1, the emitter voltage of TR1 will rise by several millivolts in relation to the voltage at the junction of R5 and R6, and the meter will read.

Full scale sensitivity of the null voltmeter is around 150 millivolts. Metal Locator response is shown in Fig. 2, where meter reading is plotted against depth for three weights of metal.

## CONSTRUCTION

Commence construction by cutting a piece of 0.1 inch matrix plain perforated circuit board to a size of $3 \cdot 1$ by 1.4 inches, and drill holes to take C4, VR1, and S1 (see Fig. 3).

Cut two brackets from a length of $1_{2}$ inch aluminium angle and drill to accept the meter terminal screws and 6B.A. circuit board mounting screws.

Bolt the brackets to the circuit board, complete with solder tags, and insert all terminal pins in the positions shown in Fig. 3.

With C4, VR1, and S1 in place on the circuit board, proceed to mount and solder the remaining components in the following order; resistors, capacitors, wire links and leads, diodes and the transistor, using a heat shunt to protect the diodes and transistors while soldering them.

Obtain a plastic beaker with lid (of minimum dimensions 5 inches high by $2^{1}{ }_{2}$ inches diameter) and cut away the centre of the lid to accept the meter MEI. Next, drill holes in the beaker for L1 leads, woodscrews, and to allow access to the circuit board controls, see Fig. 4.

When following the step-by-step instructions in Fig. 5, for making up the search coil L1, ensure that the pile windings can slide easily off the 5 inch diameter former. Short strips of insulating tape, placed sticky side out around the former, will hold the turns together and facilitate removal of the coil. Do not use Sellotape for this purpose as it is likely to damage the wire.

The metal locator frame (Fig. 4 and 6) consists of a chipboard or plywood handle, a $5_{8}$ inch diameter dowel pole, and two s.r.b.p. or Perspex sheets for the search head. Screw and glue the handle to the pole and then glue the other end of the pole to the search head top board, this assembly can then be painted.

To complete the construction, screw the


Fig. 1. Circuit diagram of the Metal Locator. The search coil L1 is mounted in the locator head and the dotted lines are the connecting wires to the circuitry.


Fig. 2. Response curves of the Metal Locator.
plastic beaker to the pole opposite the handle, securely clamp the search coil between the boards, run twin leads from L1 to the beaker, and position the battery.

In the prototype, the battery was held in place behind the meter with a rubber band, as shown in the photograph, but it could equally well be fixed inside the beaker with a small clip or elastic band.

## SETTING UP

Adjust VR1 to mid track, C4 to minimum capacitance (unscrewed), and switch on. The meter pointer should go beyond full scale. With the search coil well away from metal objects, screw in C4 until the meter reads somewhere between zero and full scale. Trim for a zero reading with VR1.

## OPERATING LICENCE

The Metal Locator described in this article is designed to operate in the frequency band specified by the Ministry of Post and Telecommunications ( 16 to 150 kHz ). The circuit design of the locator should not be altered in any way that may affect the operating frequency.

A licence must be obtained before using the locator; this costs 75p for 5 years. An application form for a licence is obtainable from the Ministry of Post and Telecommunications, Waterloo Bridge House, Waterloo Road, London, S.E.1.

If the meter fails to read, or no response is obtained from adjustment of C 4 , check for wiring errors.

A certain amount of drift will be evident immediately after the locator has been switched on, therefore allow the circuit to settle down and then readjust C4 and VR1. Locator response can then be checked with metal weights and compared with Fig. 2.

Increased sensitivity can be achieved by reducing the value of C 3 to $0 \cdot 15 \mu \mathrm{~F}$, but this will enhance circuit drift to the point where frequent adjustment of VR1 is necessary. Conversely, drift and sensitivity will be reduced if C3 is increased in value.

$3 / 4$ THICK CHIPBOARD OR PLYWOOD handle screwed and glued to pole

Fig. 4. Construction and assembly of the beaker and handle.


Fig. 5 (a). Coil wound on former (b) Coil removed from former and bound with tape (c) Shaped coil.


## Components ....

Resistors

| R1 | $100 \mathrm{k} \Omega$ |
| :--- | :--- |
| R2 | $1 \cdot 2 \mathrm{k} \Omega$ |
| R3 | $470 \Omega$ |
| R4 | $470 \Omega$ |
| R5 | $1 \mathrm{k} \Omega$ |
| R6 | $1 \mathrm{k} \Omega$ |
| All $\pm 10 \%$ | SEE |
| watt carbon. |  |

## Capacitors

C1 $0.22 \mu \mathrm{~F}$ polyester 250 V
C2 $5,600 \mathrm{pF}$ polystyrene
C3 $0.22 \mu \mathrm{~F}$ polyester 250 V
C4 500 pF mica compression trimmer
C5 $1 \mu \mathrm{~F}$ elect. 12 V

## Semiconductors

TR1 BC108 silicon npn
D1 OA81
D2 OA81

## Meter

ME1 $50 \mu \mathrm{~A}$ f.s.d. moving coil. SEW type MR 38P

## Switch

S1 S.P.S.T. sub-miniature toggle
Miscellaneous
VR1 $10 \mathrm{k} \Omega$ miniature carbon T.V. type preset B1 PP3 battery. Circuit board 3.1 inch by 1.4 inch plain, perforated 0.1 inch matrix Veroboard and Veropins. 26 s.w.g. cotton covered or enamelled copper wire, plastic beaker (see text), connecting wire, wood and screws for assembly, $\frac{1}{2}$ in aluminium angle for brackets.

## USE

The locator is now ready for use and can be used for beachcombing or searching the back garden or waste ground. The locator may be subjected to damp and the pole, in particular, should be painted for protection if nothing else.

Photograph showing the construction of the circuit board and meter mounted on the beaker lid.


Continued from page 361


Fig. 8. Approximate output frequency for various control settings.

## FINAL ASSEMBLY

Final assembly amounts to attaching the front panel to the box frame with self tapping screws, fitting the battery inside and fitting rear panel.

The generator can be connected to the input of any amplifier but the signal output level should be adjusted in accordance with that required by the amplifier input. To comply with the calibration chart given in Fig. 8 turn VRI fully anti-clockwise and fix the frequency control knob to read zero. The output control knob is fixed in the same way i.e., to read zero with VR2 fully anti-clockwise.

The Audio Tone Generator is now ready for use and can be tried out in conjunction with a tape recorder.


BRAND NEW GUARANTEED

LARGEST SELECTION OF SEMICONDUCTORS COMPONENTS

| TRANSISTORS |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 29301 | 20p | 2N3404 | 32 ${ }^{\text {p }}$ | 40310 | 45p | BC212L | 13p | BSX 28 | 32／p | NKT281 | p |
| $2 \mathrm{G302}$ | 20 D | 2N3405 | 45p | 40311 | 85p | BCY30 | 271p | Bsx 60 | 82 p | NKT401 | $871^{\circ}$ |
| 2 C 303 | 20p | 2N3414 | 22\}p | 40312 | 4710 | BCY31 | 30p | BSX 61 | 62¢p | NKT402 | 90 p |
| 2 C 306 | 421p | 2N3415 | 22．p | 40314 | 3710 | BCY32 | 60p | B8X 76 | $22 \pm \bar{\square}$ | NKT403 | 75p |
| 20308 | ${ }^{30 \mathrm{D}}$ | 2N3416 | 371． | 40320 | 471p | BCY33 | 25p | BSX 77 | 271p | NKT404 | 82／b |
| 2G309 | ${ }^{30 p}$ | 2N3417 | 371 p | 40323 | $32 / \mathrm{p}$ | BCY 34 | 30 p | BEX78 | 27\％ | NKT405 | 75p |
| $2 \mathrm{C371}$ | 15p | 2N3570 | 21．25 | 40324 | 4710 | BCY38 | 40p | BSY10 | 271p | NK T406 | 82\％${ }^{\text {D }}$ |
| 20374 | 20p | 2N3572 | 97\％p | 40326 | 3710 | BCY 39 | 60p | BSY11 | $271 p$ | NKT451 | 82 p $^{\text {p }}$ |
| 29381 | 2210 | 2N3605 | 27 bp | 40329 | 30p | BCY40 | 50 p | BSY24 | 15p | NXT452 | 62 ${ }^{\text {p }}$ |
| 2N404 | 22 b | 2N3606 | 2710 | 40344 | 2710 | BCY42 | 15 y | BSY25 | 15D | NKT453 | 47\％ |
| 2N696 | 20p | 2N3607 | 22］ p | 40347 | 67 p | BCY 43 | 15p | BSY26 | 17¢p | NKT603F | F32 ${ }^{\text {p }}$ |
| 2N697 | 17p | 2N3702 | 11 | 40348 | 681 p | BCY54 | 32 ${ }^{\text {p }}$ | BSY27 | 171p | NKT613F | F 32 ${ }^{\text {d }}$ |
| 2N698 | 25p | 2N3703 | 10p | 40360 | 42］${ }^{\text {d }}$ | BCY58 | 22.0 | B8Y28 | 171p | NKT674F | $\mathrm{F}^{30 \mathrm{p}}$ |
| 2N706 | 124p | 2N3704 | 11 p | 40361 | 4719 | BCY59 | 22］p | B8Y29 | 17．p | NKT677F | F 30p |
| 2N705A | 12\％p | 2N3705 | 10p | 40382 | $87 / \mathrm{D}$ | BCY 60 | 974p | BSY 32 | 25p | NKT713 | 25p |
| 2N708 | 15p | 2N3706 | 09D | 40370 | 3210 | BCY70 | 20p | BSY 36 | 25p | NKT781 |  |
| 2N709 | 62 p | 2N3707 | 11. | 40406 | 671 ${ }^{\text {d }}$ | BCY 71 | 25p | B8Y37 | 25p | NKT10419 | 19 30p |
| 2N718 | 25p | 2N3708 | 07p | 40407 | 40p | BCY72 | 171p | B8Y 38 | $22 \ddagger p$ | NKT1043 |  |
| 2N725 | 30p | 2N3709 | 09p | 40408 | 32］p | BCz10 | 27 1p | B8 Y 39 | 22 ${ }^{\text {p }}$ |  | 371p |
| 2N727 | 30p | 2N3710 | 09p | 40410 | 827 | BCZ11 | 421p | BSY40 | 32 p | NKT1051 | 19 |
| 2N914 | 171p | 2N3711 | 12p | 40467A | 67］p | BD116 | 21－12 | B8Y51 | 32¢p |  | 32 ${ }^{\text {p }}$ |
| $3 \times 916$ | 171p | 2N3715 | 21.25 | 40468A | 35p | BD121 | 85D | BSY52 | $32 \cdot \mathrm{p}$ | 203 |  |
| 2N918 | 30p | 2N3716 | 11.30 | 40800 | ${ }^{57}$ P | BD123 | 82\}p | 788Y53 | 37 p |  |  |
| 2N929 | 224p | 2N3791 | 82.08 | AC 107 | 30 p | BD129 | 60p | BSY54 | 40 p | NKT20338 |  |
| 2N930 | 271p | 2N3819 | 35p | AC128 | 20 p | BD131 | 75p | BEY56 | 80 p |  | 87 ｜D |
| 2N1090 | 22］p | 2 N 3823 | 9710 | AC127 | 25p | BD132 | 85p | BEY78 | 474 ${ }^{\text {d }}$ | NKT80111 |  |
| 2N1091 | 22，${ }^{\text {d }}$ | 2N3854 | 2710 | AC128 | 200 | RDY10 | 21．371 | BSY79 | 45p |  | 77 |
| 2N1131 | 25D | 2 N 3854 A | ${ }^{271 p}$ | AC154 | 221 p | BDY11 | ${ }^{11.624}$ | B8Y82 | 58\％p | KT8011 |  |
| 2 N 1132 | 25p | 2N8955 | $271 p$ | AC176 | 25p | BDY17 | 81.50 | BSY90 | 57 ${ }^{\text {p }}$ |  | 97ip |
| 2N1302 | $17 \%$ | 2N 3855 A | 30p | AC187 | 621p | BDY18 | 81.75 | BSY90A | 12¢p | NKT8011 |  |
| 2 N 1303 | 17tp | 2N3856 | 30 D | AC188 | 8710 | BDY19 | ع1．97 | B8w 11 | 42 |  | 11.12 |
| 2 N1304 | 22tp | 2N3856A | A 35p | ACY17 | 2710 | BDY20 | \＆1－12 | Bsw70 | 27 ip | NKT8021 |  |
| 9N1305 | 22 p | 2N3858 | 25 p | ACY18 | ${ }^{25 p}$ | BDYs8 | 97tp | ${ }^{\text {C111 }}$ | 75p |  |  |
| 2N1306 | 25D | 2 N 3858 A | A 30p | ACY19 | 25 p | BDY 6 | 21.25 | C424 | 27\％${ }^{\text {b }}$ | T8021 |  |
| 2N1307 | 25 D | 2N3858 | 27b | ACY20 | ${ }^{25 p}$ | BDY61 | £1．25 | C425 | 55 p |  | 82tp |
| 2N1308 | 30D | 2 N 3859 A | 32］ | ACY21 | 250 | BDY62 | 21.00 | C426 | 40p | NKT8021 |  |
| 2N1309 | 30p | 2N3860 | ${ }^{30 p}$ | ACY22 | 20p | BF115 | 25p | C428 | 3710 |  |  |
| 2N1507 | 1710 | 2N3866 | \＄1．50 | ACY 28 | 20 p | BF117 | 47¢ ${ }^{\text {d }}$ | C744 | ${ }^{30 \mathrm{D}}$ |  |  |
| 2N1613 | ${ }^{2.5 p}$ | 2N3877 | 40p | ACY 40 | 20 p | BF163 | 37］p | D16P1 | 37.1 |  |  |
| 2N1631 | 350 | 2N3877A | 40 D | ACY 41 | 25p | BF167 | 18p | D16P2 | 40 p | NKT8021 |  |
| 2N1632 | 30D | 2N3900 | 3710 | ACY44 | 40p | BF173 | 19p | D16P3 | 3710 |  | 82ip |
| 2N1638 | 2710 | 2 N 3900 A | 40p | ADI40 | $52 \ddagger$ | BF177 | ${ }^{30 p}$ | D16P4 | 40 p | NKT8021 |  |
| 2N1639 |  | 2N3901 | 97 fp | AD149 | 67\％${ }^{\text {P }}$ | ${ }^{\text {BF178 }}$ | ${ }^{30 p}$ | GET102 | ${ }^{30 D}$ |  | 21p |
| 2N1671B | 11．00 | 2N3903 | 35p | AD150 | 621p | BF179 | ${ }^{30} \mathrm{p}$ | GET113 | 20p | 20 | 75p |
| 2N1711 | 25p | 2N3904 | 35p | AD161 | 371p | BF180 | 35 p | GET114 | 20 D | $\mathrm{OC2}^{2}$ | 0 p |
| 2N1889 | 3210 | 2 N 3005 | 37.18 | AD162 | 37 \＄D | BF181 | 32 h | GET118 | 20 p | 0 C 23 | ${ }^{60 p}$ |
| 2N1893 | 37.10 | 2 N 390 B | 37 pp | AF106 | 42 ${ }^{2} \mathrm{P}$ | BF184 | 25 p | GET119 | 20 D | ${ }^{\text {OC2 }}$ | ${ }^{60 p}$ |
| 2N2147 | 8810 | 2N4058 | 17\％${ }^{\text {d }}$ | AF114 | 25 D | BF185 | 421 p | GET120 | 521p | ${ }_{0}^{0} 23$ | 50 p |
| 2N2148 | 8710 | 2N4059 | 10p | AF115 | 25 p | BF194 | 17］p | GET873 | 12！${ }^{\text {d }}$ | ${ }^{0} \mathbf{0} 26$ | 971 |
| 2N2160 | 3710 | 2N 4060 | 12\％p | AF116 | 250 | BF195 | 15 p | GET880 | ${ }^{30 p}$ | $\mathrm{OCP}^{8} 8$ | 820 |
| 2N2193 | 10p | 2N4061 | 12¢p | AF17 | 25p | BF196 | 421 p | GET887 | 20 D | 0029 | 821 p |
| 2N2193A | 421p | 2N4062 | 221p | AF118 | 62 ¢ ${ }^{\text {d }}$ | BF197 | 42 ¢ | GET889 | 221 p | OC35 |  |
| 2N2194A | 30p | 2N4244 | 47\％${ }^{\text {d }}$ | AF119 | 200 | BF198 | 42］p | GET890 | 2210 | ${ }^{0} \mathrm{C} 36$ | 62ヶp |
| 2N2217 | 27p | 2N4285 | 171p | A F124 | 22.1 | BF200 | 521 p | GET896 | 221 | 0C41 | 22 p |
| 2N2218 | 23p | 2N4286 | 171p | AF125 | 20 p | BF 224 | 14 p | GET897 | 22tp | OC42 | 2p |
| 2N2219 | 23p | 2N4287 | 17\％p | AF126 | 20p | BF225 | 19p | GET889 | 2219 | OC44 | 20p |
| 2N2220 | 250 | 2N 4288 | 171p | AF127 | 171p | BF237 | 23p | MJ400 | 21．071 | 0 C 45 | 12\％p |
| 2N2221 | 25p | 2N4289 | 171p | AF139 | 37 ¢p | BF238 | 23p | MJ 420 | 21.181 | $\mathrm{OC}^{\text {C }} 6$ | ${ }^{5 p}$ |
| 2N2222 | 30 D | 2N4290 | 1710 | AF178 | 4210 | BF244 | 23p | MJ421 | E1．121 | 0 C 70 | 15p |
| 2N2270 | 4710 | 2N4291 | 17 ${ }^{\text {p }}$ p | AF179 | 72 p | BFW81 | 471 p | MJ430 | 81.021 | 0 C 71 | 12ip |
| 2N2297 | 30p | 2N4292 | 181p | AF180 | $52 \%$ D | BFX12 | 2210 | MJJ40 | 95p | OC72 | $12 \ddagger p$ |
| 2N2368 | 1710 | 2N4303 | 47 LD | AF181 | 42 p | BFX13 | $22+\mathrm{p}$ | MJ480 | 8718 | ${ }^{0} \mathrm{CO} 75$ | ${ }^{32}+\mathrm{p}$ |
| 2N2369 | 17 10 | 2N5027 | 52 p | AF239 | 42\％ | BFX29 | ${ }^{30 \mathrm{D}}$ | MJ481 | 21.25 | ${ }^{0} 785$ | 224 p |
| 2 N 2369 A | 171p | 2N5028 | 57／p | AF279 | 47 ¢p | BFX 30 | 300 | M．4490 | ¢1．00 | ${ }^{0} \mathrm{C} 76$ | 22！p |
| 2 N 2410 | 421p | 2N5029 | 471］ | AP280 | 62 D | BFX42 | 8710 | MJ491 | 21．37 | $0 \mathrm{C77}$ | 30 p |
| 2N2483 | 27¢p | 2N5030 | 42 p | AF2Il | 32， p | BFX44 | 871 p | MJ1800 2 | 28．171 | 0 c 81 | 20p |
| 2N2484 | 32 ${ }^{\text {p }}$ | 2N5172 | 12．p | A8Y26 | 26p | BFX 68 | 871］ | MJE340 | 62t． | $0 \mathrm{C81}$ | 22ヶp |
| 2N2539 | 22tp | 2N5174 | $52 . p$ | A8Y27 | 371 P | BFX84 | 25p | MJES20 | ${ }^{80 \mathrm{D}}$ | $0 \mathrm{OC83}$ | 25p |
| 2N2540 | 22¢p | 2N5175 | $52+\mathrm{p}$ | A8Y28 | $27 \%$ | BFX85 | 3210 | MJE521 | 73p | OC84 | 25p |
| 2N2613 | 35p | 2N5178 | 45p | A8Y29 | 27 ¢p | BFX86 | ${ }^{25}$ | MPF102 | 422 ${ }^{\text {2 }}$ | OC139 | ${ }^{321 p}$ |
| 2N2614 | 30p | 2 N 5232 A | A 30p | Asy 36 | 25 D | BFX87 | 2710 | MPF103 | 37 p | $0 \mathrm{Cl140}$ | 321p |
| 2N2646 | 62 ${ }^{\text {d }}$ | 2N5245 | 45p | A8Y50 | 265 | BFX88 | 25p | MPF104 | ${ }^{371 p}$ | OC170 | 30 p |
| 2N2696 | $32 \cdot \mathrm{p}$ | 2N5246 | 42］p | ASY51 | 321 D | BFX89 | 62.5 | MPF105 | 37 p | 0 O 171 | 30 p |
| ${ }^{2 N} \mathbf{2 N 2 7 1 1}$ | ${ }_{250}^{25 p}$ | ${ }_{2 \text { N } 5285}$ |  | A8Y54 |  | ${ }_{\text {BFX }}{ }^{\text {Bra }}$ |  |  | 32 ${ }^{\text {P1 }}$ |  |  |
| $2 \mathrm{~N}_{2}^{2713}$ | ${ }_{2710}^{250}$ | ${ }_{\text {2N6268 }}^{\text {2N5 }}$ | ${ }_{82.75}^{23.25}$ | A8Y86 | 32．${ }^{3}$ | BFY ${ }^{\text {BFI }}$ | $321 p$ $42 p$ | NKT001 | 429p | －C202 | ${ }^{60 p}$ |
| 2N2714 | 309 | 2N6287 | £2．62t | A8Z21 | 481 D | BFY17 | 2210 | NKT125 | 2710 | OC203 | $42+\mathrm{D}$ |
| 2N2865 | 62¢p | 2N5305 | 374 p | BC107 | 10 p | BFY18 | 32 p | NKT126 | 275 | ${ }^{\text {OC204 }}$ | 2， $\mathrm{p}^{\text {p }}$ |
| 2N2904 | 30p | 2N5306 | 400 | BC108 | 10 D | BFY19 | 8210 | NKT128 | 27ip | OC205 | 0 p |
| $2{ }^{2} 2904 A$ | 32 ${ }^{\text {p }}$ | 2N5307 | 37 p | BC109 | 10 D | BFY20 | 21.60 | NKT135 | 27 p | $\mathrm{OCO}^{\text {O }}$ O7 | 75 p |
| 2N2905 | 3710 | 2N6308 | 371 D | BC113 | 150 | BFY21 | 42 p | NKT137 | 32 p | ocre7 | 42：p |
| 2N2905A | 40p | 2N5309 | 621 p | BC115 | 15 p | BFY24 | ${ }^{45 p}$ | NKT210 | 30 p | ORP12 | 50p |
| 2 N 2906 | 25p | 2N5310 | $4{ }^{4} \frac{1}{}{ }^{\text {D }}$ D | BC118A | 150 | BFY25 | 25p | NKT211 | 300 | ORP81 | 50p |
| 2N2906 | 27p | 2N5354 | 27.1 | BC118 | 100 | BFY26 | 20 p | NKT212 | 30 p | P346A | 221 D |
| 2N2907 | 30p | 2N5355 | $27 . \mathrm{P}$ | BG121 | 20 p | BFY29 | 50 p | NKT213 | 30p | T1834 | $62{ }^{\text {\％}}$ |
| 2N2923 | 15p | ${ }_{2}^{2 N 5356}$ | 32.1 | $\mathrm{BCl2}^{\text {a }}$ | ${ }^{200}$ | BFY30 | 500 | NKT214 | ${ }^{224}$ | TIS43 | ${ }^{279}$ |
| 2N2924 | $15 p$ 150 | ${ }_{2}^{2 N 5365}$ | ${ }^{47}$ p |  | 200 | ${ }_{\text {BFY }}{ }^{\text {BFY }} 4$ | \％${ }_{60 \mathrm{p}}$ | ${ }_{\text {NKT218 }}$ | 294p | T1844 T1845 | 109 109 |
| 2N2928 | 150 | ${ }^{2 N 63687}$ | ${ }^{87 \%}$ | $\underset{\text { BC140 }}{ }$ | ${ }_{370}$ | BFY ${ }^{\text {BFY }}$ |  | NKT218 | ${ }_{\text {42，}}{ }^{\text {d }}$ | T1846 | 110p |
| Green | 14p | 2N5457 | 3710 | BC147 | 10 p | BFY51 | 2 n | NKT219 | ${ }^{30 p}$ | T1847 | 12 p |
| Yeilow | 12．p | 25005 | 75 p | BC148 | 10p | BFY52 | 23p | NKT223 | 27 P | T1848 | 124 p |
| Orang | 12 ${ }^{\text {p }}$ | 28020 | 22．00 | BC149 | 12p | BFY53 | 171p | NKT224 | 20， | T1849 | 12 p |
| 2N3011 | 30 p | ${ }_{28103}$ | ${ }^{50 \mathrm{p}}$ | $\mathrm{BCL}^{2}$ | 1710 | BFY56A | 67.5 | NKT225 | 221P | TI850 | $17 / \mathrm{p}$ |
| 2N3014 | ${ }^{32} 18$ | ${ }^{28103}$ | ${ }^{25}$ | ${ }_{\text {BCl }}{ }^{\text {B7 }}$ | ${ }^{20 p}$ | ${ }_{\text {BFY75 }}$ | ${ }_{4}^{30 p}$ | NKT229 | ${ }_{35}^{301}$ | T1831 | 12p |
| 2N3053 | 18 p | 28104 | 250 | BC158 | 11 p | BFY78 | 42¢p | NKT237 | 35 p | T1832 | 12．p |
| 2N3054 | 48p | 28501 | 32 j | BC159 | 12p | BFY77 | 57p | NKT238 | 25p | TI853 | 29 ［ |
| 2N3055 | ${ }^{\text {a2p }}$ | ${ }_{2}^{28502}$ | ${ }^{35 \mathrm{p}}$ | BC160 | 62.1 | BFY90 | ${ }^{67 \%}$ | NKT240 | 271 p | TI860 | 224 |
| 2 N 3133 | 30 p | ${ }^{28503}$ | 2710 | ${ }^{\mathrm{BC} 167}$ | 11. | RFW58 | 27.1 | NKT241 | ${ }^{271 p}$ | ${ }_{\text {T1882 }} 181$ | ${ }^{251}$ |
| 2N3134 | 30 p | 3N83 | 40 p | BC168B | 100 | BFW69 | 25p | NKT242 | ${ }^{20 p}$ | T1882 | 2710 |
| 2N3135 | 25 p | 3N128 | 70p | RC188C | 11p | RFW60 | 25 p | NKT243 | ${ }^{624}{ }^{17}$ | T1P29A | 50 p |
| 2N3136 | 25p | 3N140 | 7710 | BC169B | 11. | BPX25 | 81.85 | NKT244 | 17id | Tip30a | ${ }^{60 p}$ |
| 2N3390 | 25 p | 3N141 | 72.5 | ${ }^{\text {BCl } 690}$ | 12p | BPX29 | ${ }^{21-80}$ | NKT245 | 20p | TIP31A | ${ }^{621 p}$ |
| 2N\＄391 | 20 p | 3N142 | 55p | BC170 | 12.0 | BPY10 | 81.45 | NKT261 | ${ }^{20 p}$ | TIP32A | 759 |
| ${ }^{2 N 3391 A}$ | 30p | 3N143 | ${ }^{87} 7$ | BC171 | 15 p | Bry39 | 37ip | NKT282 | 30 p | TIP33A |  |
| 2N3392 | 17ip | 3N152 | 87.1 | BC172 | 100 | B8x19 | 17p | NKT264 | 200 |  | 02］p |
| 2N33893 | ${ }^{15 p}$ | R．C．A． | 52pp | ${ }_{\text {BC182 }}$ | $226 p$ $10 p$ | B8X20 | $171 p$ 3760 | NKT271 | 20 p | TIP35A | 82.05 $£ 2.90$ |
|  | 5 | 40251 |  | ${ }_{\text {BC182 }}$ | ${ }_{09 p}$ | B8X26 | $4{ }^{4} 5$ | NKT274 | ${ }_{80 \mathrm{p}}$ | TIPソ6A |  |
|  | $22 \mid$ | 40309 | 32 \＄ | BCI84 | 11p | X27 | 471p | NKT275 | 20p |  |  |
| Post \＆Packing 13p per order． Matehing eharge（audio |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |

## TTL．LOGIC I．C．NEW PRICES

|  | 1－11 12－24 |  |  | 1－11 12－24 |  |  | 1－11 12－24 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Ep | Ep |  | ip | 2p |  | Ep | \＆p |
| SN7400 | 0.20 | 0.18 | SN7433 | 0.80 | 0.75 | 8N7472 | 0.32 | 0.30 |
| gN7401 | 0.20 | 0.18 | 8N7437 | 0.84 | 0.08 | SN7473 | 0.43 | 0.41 |
| SN7402 | 0－20 | 0.18 | gN7438 | 0.84 | 060 | 8N7 ${ }^{\text {7 }}$（ | 0.43 | 0.41 |
| SN7403 | 0.80 | 0.18 | SN7440 | 0.23 | 0.21 | SN7475 | 0.45 | 0.44 |
| SN7405 | 0.20 | 0.18 | SN7441AN | 0.87 | 0.83 | 8N7476 | 0.45 | 0.41 |
| gN7406 | 0.80 | 0.75 | 8N7442 | 0.85 | 0.81 | EN7480 | $0 \cdot 70$ | 065 |
| SN 7407 | 0.80 | 0.75 | EN7443 | 2.86 | $2 \cdot 70$ | SN7481 | 1.40 | $1 \cdot 38$ |
| SN7408 | 0.20 | 0.18 | gN7444 | 2.86 | 2－70 | 8N7482 | 0.87 | 082 |
| SN7409 | 020 | 0.18 | 8N7445 | 2.50 | $2 \cdot 40$ | SN7483 | 0.87 | 082 |
| 8N7410 | 0.20 | $0 \cdot 18$ | gN7446 | 1.00 | 0.95 | SN7484 | 2.00 | 1.85 |
| 8N7411 | 023 | 021 | 8N7447 | 1.00 | 0.95 | SN7480 | $3 \cdot 62$ | 3.40 |
| gN7412 | 0.48 | 0.48 | gNT448 | 1.00 | 0.95 | 8N7486 | 0.33 | 030 |
| 8N7413 | 0.40 | 0.38 | SN7449 | 1.00 | 0.85 | SN7430 | 0.87 | 0.84 |
| gN7420 | 0.80 | 0.18 | SN7450 | 0.20 | 0.18 | 8N7491AN | 1.21 | $1 \cdot 10$ |
| \＄N7423 | 0.51 | 0.47 | 8N7451 | 0.20 | 0.18 | SN7492 | 0.87 | 0.81 |
| gN7427 | 0.48 | 0.45 | BN7453 | 0.20 | 0－18 | SN7493 | 0.87 | 0.84 |
| 8N7428 | 0.80 | 0.75 | SN7454 | 020 | 0.18 | SN7494 | 0.87 | 0.84 |
| ¢N7430 | 0.23 | 0.15 | SN7460 | 0.20 | 0.18 | 8N7495 | 0.87 | 0.84 |
| 8N7432 | 0.48 | 0.42 | SN7470 | 040 | 0.38 | SN7496 | 0.87 | 0.84 |

MULLARD SUB－MIN ELECTROLYTIC
 $6 \cdot 4 / 6 \cdot 4 ; 6 \cdot 4 / 25 ; 8140 ; 10 / 1$ A；10／64； $12 \cdot 5 / 25 ; 16 / 40 ; 20 / 16 ; 20 / 64 ; 25 / 6 \cdot 4$ 25／25； $32 / 10 ; 32 / 40 ; 32 / 64 ; 40 / 16 ; 50 / 8 \cdot 4 ; 50 / 25 ; 50 / 40 ; 64 / 10 ; 80 / 2 \cdot 6$ 80／16：80／25；100／6．4；125／10；125／16；200／10；320／6．4．

## SILICON RECTIFIERS

| PIV | 50 | 100 | 200 | 400 | 600 | 800 | 1000 | 1200 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 A | 8 p | 9p | 10p | 13 p | 12p | 15p | 20p | － |
| 3 A | 15p |  |  | 22ip |  | 30 p |  |  |
| ${ }^{64}$ |  |  | 25 p | ${ }^{30 p}$ | 32 p | 35 p |  |  |
| 10A | 30p | 35p | 40p | 47D | 56 D | 66 p | 75p |  |
| 15A | ${ }^{36 \mathrm{p}}$ | 45 p | 48 D | 55p | ${ }^{\text {asp }}$ | 75p | 87 p |  |
| $35 \mathrm{~A} \quad 70$ |  | 80 p | 90 p | £1．00 | £1．40 | ¢1－70 | 22．76 |  |
| 1 amp | 3 an | are p | tie e | prulat |  |  |  |  |

DIODES \＆RECTIFIERS

| IN34A | 10p | AA119 | 7p | BAX16 | 121p | P9T3／4 | 22pp |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IN914 | 70 | AA129 | 15p | BAY18 | 178 | OAS | 17p |
| IN916 | 7p | AAZ13 | 12 p | BAY31 | 7p | OA10 | 20p |
| IN4007 | 20p | AA715 | 12p | BAY38 | 25 D | OA， | 10p |
| IS44 | 7p | AAZ17 | 10p | BY100 | 15p | OA47 | 8 |
| 18113 | 15p | BA100 | 15p | BY103 | 22 D | OA70 | 7p |
| 18120 | 12p | baloz | 25p | BY122 | 471p | OA73 | 10p |
| 18121 | 14 D | BA110 | 25p | BY124 | 15p | OA79 | 7p |
| 18130 | 8 p | BA114 | 15p | BY128 | 15p | OA81 | 8 |
| IS131 | 10D | BA115 | 7p | BY127 | 17p | OA85 | 10 p |
| Is132 | 12p | BA141 | 170 | BY164 | 57 | OA90 | 7p |
| 18920 | 7 D | BA142 | 17p | BYX 10 | 22p | OA91 | 7p |
| 18922 | 8 D | BA144 | 12p | BYZ10 | 35 | 0495 | 7 |
| 18923 | 12p | BA145 | 17p | BYZ11 | 32p | OA200 | 7p |
| 18940 | $5{ }^{5}$ | BA154 | 12p | $1 \mathrm{YYZ2}$ | 30 p | OA202 | 10 p |
|  |  | BAX 13 | 5p | HYz13 | 25p | TIV307 | 50p |

＂SCORPIO＂CAP
DISCHARGE IGNITION
SYSTEM
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## TREABHIN Half term test ATSMTHR <br> 

Last month we posed some problems under the heading Teach in Half-Term Test. We will now answer those problems and try to show how we arrived at the answers. If you have got some of them wrong do not worry, just try and follow our explanation and see where you went wrong.
(1) They flow from negative to positive in reality. Although we assume that conventional current flows from positive to negative the actual electrons flow from negative to positive.
(2) (b) $\mu \mathrm{A}($ microamps), (e) $\mathrm{A}(\mathrm{amps})$
(3) 22 volts. $V=I R$ hence $V=0.01 \times 2.2 \times 1,000=22 V$
(4) It does not matter. All the resistor does is to limit the current; this can be done at any point around the circuit.
(5) 2.8 mA . Total resistance is $2.2 \mathrm{k} \Omega+1 \mathrm{k} \Omega=3.2 \mathrm{k} \Omega$. Current flow $\mathrm{I}=\frac{\mathrm{V}}{\mathrm{R}}=\frac{9}{3.2 \times 1,000}=2.8 \div 1,000 \mathrm{~A}=2.8 \mathrm{~m} \mathrm{~A}$
(6) R1 and R3 $\frac{1}{2}$ W, R2 1 W . Total circuit resistance $R_{T}=R 1+\frac{R 2 \times R 3}{R 2+\bar{R} 3}=10+33 \cdot 3=43 \cdot 3 \Omega$.
Total current $I \Rightarrow \frac{V}{R}=\frac{9}{43.3}=0.21 \mathrm{~A}$
Dissipation of $R 1=1^{2} R=0.21 \times 0.21 \times 10=0.44 \mathrm{~W}$. The nearest commercial rating is $\frac{1}{2} W$. Next calculate the voltage drop across $R 2$ and $R 3$ together $V=I R=0.21 x$. $33 \cdot 3=7 \mathrm{~V}$.

We know that $W=I^{2} R$, but $I=\frac{V}{R}$ therefore
$W=\frac{V}{R} \times \frac{V}{R} \times R$ and, cancelling $W=\frac{V^{2}}{R}$
Dissipation in $\mathbf{R} 2=\frac{\mathrm{V}^{2}}{\mathbf{R}}=\frac{7 \times 7}{50}=\frac{49}{50}=0.98 \mathrm{~W}$
Dissipation in R3 $=\frac{\mathrm{V}^{2}}{\mathrm{R}}=\frac{49}{100}=0.49 \mathrm{~W}$
(7) 0.4 W or 400 mW . Maximum dissipation occurs when the value of VR1 equals that of R1 i.e. 50S. When both resistors are of equal value the voltage drop across each is half the voltage drop across both, therefore, maximum dissipation in VR1

$$
=\frac{V^{2}}{R}=\frac{4.5 \times 4.5}{50}=\frac{20.25}{50}=0.405 \mathrm{~W}
$$

(8) (a) $4 \cdot 7 \mathrm{k} \Omega \pm 10 \%$
(b) $22 \mathrm{k} \Omega \pm 5 \%$
(c) $100 \mathrm{k} \Omega \pm 10 \%$
(9) (b) $20 \mu \mathrm{~F} 40 \mathrm{~V}$. In most applications using electrolytic capacitors the capacitance must be greater than a certain value; the tolerance of a normal $16 \mu \mathrm{~F}$ would encompass $.20 \mu \mathrm{~F}$. The important thing is that the working voltage is the same or greater.
(10) Reject it politely. He has given you a $120,000 \mathrm{pF}$ or $0.12 \mu \mathrm{~F}$ capacitor. Check to see If he has the precise value and, if he does not, you may as well take this one, since it should be near enough to use as a substitute.
(11) C1 will charge up the fastest as it has the lowest value and is being charged through the lowest value resistor.
(12) C2 will take the longest time to charge, as it has the highest value and is being charged through the highest value resistor.
(13) Forward biassed. The conventional current flows from positive to negative and can thus flow through the diode in the direction of the arrow.
(14) 100 V and 100 mA . Peak reverse breakdown voltage will be the battery voltage. Since in the reversed blassed condition there is negligible current flowing R1 will not drop any voltage and the full supply voltage will appear across D1. In the forward biassed condition the diode can be assumed to be a short circuit thus only R1 can limit the current flowing hence

$$
1=\frac{V}{R}=\frac{100}{1 \times 1,000}=0.1 \mathrm{~A} \text { or } 100 \mathrm{~mA}
$$

(15) (d) $100 \mathrm{~V}, 150 \mathrm{~mA}$. Both ratings given are minimum ratings, $0.1 \mathrm{~A}=100 \mathrm{~mA}$.
(16) (b) 0.6 V . As the diode is forward biassed the voltage would be 0.6 V . There is always a voltage drop of approximately 600 mV across silicon diodes due ta the "knee" in the characteristic.
Well, how did you fare? If you got them all right that is excellent, if you did not the important thing is that you understand where you had difficulties. We suggest that you re-read the relevant sections of the Teach-In series.

We hope that you found the questions a challenge and at the same time they have opened your eyes to some calculation methods-particularly the calculation of dissipation. If you used $W=I^{2} R$ instead of deriving $W=\frac{V^{2}}{R}$ this does not matter but it may pay to look for an easier way next time.


## Bias Value

Having been a subscriber to P.E. and P.W. "off and on" for about 10 years I came across the January issue of Everyday Electronics, which had my instant approval and now joins the rank of my other magazine's culminating in an endless and very informative pile on top the piano.

I find it is a magazine not only of theoretical enthusiasm but of great practical interest to the "everyday handyman" and certain to be a book for beginners, especially the very helpful facts "projected" by Mike Hughes, M.A.

I would hope in the future that perhaps Mr. Hughes could give reference to finding values of bias resistors, etc., needed for the satisfactory operation of different transistor parameters, and also relevant circuit operation of thyristors, unijunction and field effect transistors and other very useful flexible types of semiconductors.

Noticing other readers' troubles referring to the Electro Laugh, I also constructed this article and it worked first time owing to the way I adopt when working on, or constructing any project, I always check the finished article with the actual circuit diagram thus finding our little friend Q7 and P7.

Unfortunately the only earphone I had was a high impedance crystal type, but by connecting a resistor in the region of 250 ohms in parallel with it, it brought the overall impedance down to a satisfactory level with a slight reduction in volume.
J. Mason
S. Wales

We doubt if Teach-In will be able to meet all your needs as it will finish after 12 months. However we will be publishing further series that should help.

## Another Bug

Naturally, I was quite flattered to discover that you had found my letter sufficiently interesting for inclusion in Readers Letters (March issue), however, I must admit that my pleasure was mixed with large helpings of disappointment and frustration due to your editing of the letter.

I am not complaining at all about the amount of space allocated to my comments-I realise you have the right to include only that which in your wisdom you decide is worthy of publication.

My complaint is that you have entirely neglected to make even a brief reference to what was after all the main point. of my letter-the difficulty of obtaining items advertised in your magazine. By omitting any reference to this frustrating situation, my letter as printed is sailing under false colours-the few minor constructional queries were in fact, sorted out by trial and error once I got going. The real reason for being unable to get cracking was not so much mounting components, as actually getting hold of them!

The fact that you completely ignored my comments regarding suppliers leads me to two conclusions:
(One) That you accepted my comments to be an exaggeration of a somewhat hysterical nature, and were not a true picture of the real situation, or
(Two) That you accepted my statements as correct, but did not wish to offend your advertisers whose business you must obviously wish to retain.

With regard to the former, I feel I. must now justify my remarks by quoting a few of the more deplorable examples of SERVICE, and leave you to form your own conclusions. These examples are on a separate sheet herewith enclosed.

Regarding (Two), whilst I
realise that you are not to be held responsible for goods or services advertised in your columns, you do, however, have a moral responsibility to your readers. After all, it is you that place these offers before us, the readers, and if for example, I had not seen a certain item offered in your magazine, then I would have been saved the trouble and frustration that followed when the item failed to 'arrive, and all attempts to obtain satisfaction are largely ignored.

However, I have now found a couple of very good suppliers whose friendly, courteous, and extremely efficient service have allowed me to obtain some of the pleasure that I had hoped would be derived from my new hobby (Galleon Trading Co. and Radio Exchange Co.).

To date I have completed several very efficient radios, some from kits; also the Astron, a general purpose amplifier, and one or two other gadgets, and success rate so far is quite satisfactory, so the situation is not too black after all.
J. G. Richards

Sale, Cheshire
The above correspondent supplied us with details of orders placed with four different advertisers, none of which had been expediently dealt with, at the time of writing.

We have investigated all of these cases on behalf of our reader. The delays, regrettable as they are, seem to be unavoidable and can be largely attributed to the phenomenal success of this magazine's declared intention to popularise the hobby of electronics!

As a consequence, our advertisers are sometimes overwhelmed by a flood of orders, and delays do therefore sometimes arise. But we know all our advertisers make determined efforts to clear their back-log of orders as quickly as possible.

We, on our part, will always investigate any serious and reasonable complaints, on behalf of our readers.

## Cell Life

I have just read the March issue of Everyday Electronics and thoroughly appreciated the Ruminations by Sensor where he mentioned the tin saw and how much damage could result to a \& 15,500 r.p.m. shaft is $\$ \mathrm{ln}$. dia meter 230/240v. Its speed may be further controlled with the use of our Thyrister controller. Very powerful and useful motor size approz. 2 in . dia. $\times 5$.in. long. Price $88 p$
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Made by Bmitha, these are AC mains opersted, NOT CLOCKWORK. Ideal for mounting on rack or shel or can be built into bor with 13 a socket. 2 com-
pletely adjustable pletely djustable time perlods per 24 hours, o amp changeover contacte will switch clrcult on or ofl during these periods. 50 p pair.

[^1]
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QUANTITY
OFFERS:
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beginner's enthusiasm.
When I read about the Signal Injector by Alan Jardine I was reminded of the poor beginner.

Following the instruction to solder the leads direct on to the cell will result in heating up the electrolyte and a very short life for the cell.

Perhaps this is not important as the choice of a push-on/pushoff switch allows no easy means of knowing if the thing is on or off. Very few beginners will remember to test each time, and cell life will be short it is expected. A push button perhaps?
The blind cannot be expected to lead the blind, and beginners are usually short of experience.
R. Quorn

Sussex
Of your two points concerning the Signal Injector, the first is a bit exaggerated. It is true that the cell life will be reduced by applying heat (from soldering iron) to the battery terminals but this is only negligible for the time required to execute the connection.

To install a holder to suit this type of battery would increase the cost by about 40 per cent.

We agree that it will be difficult to tell if the unit is on or off when not in use, but it can be determined; when the unit is "on" the push button will feel "loose" but in the "off" position this looseness disappears.

If this proves unsatisfactory a push-to-make release off type can be substituted.

## Wore Accurate Timer

May I thank you for publishing another article combining the hobbies of electronics and photography (ref. Darkroom Timer, March issue).

Although of excellent design, I feel it must be stated that a timer with only a 5 second timing intervals is not nearly accurate enough for the demands of the high quality black and white or well balanced colour prints that are required. However, with a small modification, I have found that the timer may be converted to an accurate piece of equipment having a timing range of 5 to 45 seconds in one second steps.

The modification requires four extra components, which are a 5 position two-pole switch (S4), VR5, VR6 and VR7 which are
skeleton presets of the values, 5 kilohm, 10 kilohm and 20 kilohm respectively.

These components form an additional timing circuit which is connected in series with the original ( $\mathbf{R}_{\mathrm{t}}$ ).
Position 1 of the switch has no further resistance and acts as a short circuit; position 2 connects VR5 into circuit, whilst position 3 connects VR6; position 4 connects VR5 and VR6 and position 5 connects VR7 into circuit.

Each position of the new switch is to represent a further one second delay.

Position 1 of course, has no further delay, position 2 however, will give a one second delay, position 3 two seconds etc. when the presets are set as they were in the original timing circuit.

Now, any time, in one second steps may be selected from 5 to 45 seconds by selecting the required 5 second range, plus the required extra time (if any) on the new switch.
D. G. Smith
Emsworth, Hants.

## Components

Let me say first of all how much I enjoy your magazine and as a newcomer to electronics I find your Teach-In articles very interesting and also Shop Talk, etc. However, I wonder if I may make a suggestion?
I constructed your Demo-Deck and find that in following this series for a month or so there is a list of the more minor components used in the experiments and I wondered if it would be at all possible, either, preferably, if you could publish the list of all the components that would be required for the rest of this series in one complete list or if possible broken up into the individual months during which they will be required.

The reason I say this is, that I,
like many of your other readers no doubt, have no local supplier of components in my immediate vicinity and it usually means a trip to Edinburgh or Glasgow to purchase these components.
However, if I could have a full list this would make things much easier for me. It would also make it much easier to send off a full list by post to a mail order firm rather than asking for two or three small components every month or so. I wonder if this could be done.

I am very grateful to you and wish you every success for your future publications.

## R. L. Grant

 ScotlandIt was our intention to publish an advanced list and in future we shall be publishing, at the end of each Teach-In every month, a list of components additional to those you have already acquired.

## Calling Gloucester

Now that I'm receiving your magazine on regular order and greatly enjoying it, I feel that I ought to go a stage further in order to get any lasting benefit from your guidance.

Can I please find out through your pages how many people in the Gloucester area are willing to ask for, and attend, an evening class on useful, basic "everyday electronics"?

Should anyone be interested, could they please write to me at the address given, then provided enough wish it, our local Education Authority can be approached with evidence that the need for such a class does exist.

Many thanks for giving me a chance to ask for these people through your very sensible magazine.
E. L. Payn

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| 3 | 4 | Photo Calls, Sun bateries. | 50p |
| :---: | :---: | :---: | :---: |
| B79 | 4 | IN\&007 Sil. Rec. diodes. 1,000 PIV lamp plastic | 50p |
| 8日1 | 10 | Reed Switehes, mixed \&ypes larite and smalf | 50p |
| -99 | 200 | Mixed Capacitors. Approx. quantity, counted by weight | 50p |
| H4 | 250 | Mixed Resintorn quancity Approx counted by | 50p |
| H7 | 40 | Wirewound Resistors. Mixed types and values. | 50p |
| H8 | 4 | ByI27 Sil. Rees. <br> 1000 PIV. 1 amp. plastic | 50p |
| н9 | 2 | OCPT Lisht Sensitive Photo Tronsiztor | 50p |
| H | 50 | NKTI55/2.59 Germ, diodes, brand new stock clearance | 50p |
| H18 | 10 | Oc71/75 uncoded black elass type PNP Gorm. | 50p |
| मां | 10 | OCBI/81D uncoded white glass type PNP Germ. | 50p |
| H28 | 20 | OC2001/2/3 PNP silicon uncoded TO. 5 can | 50p |
| H29 | 20 | OA47 Eotc bonded dioder codod MCS2 | 50p |

NEW UNMARKED UNTESTED PACKS

| 36 | 150 | Germanium Diodes | 50p |
| :---: | :---: | :---: | :---: |
| Be3 | 200 |  call croe | 50p |
| 814 | 100 |  | 50p |
| 886 | 50 | sil. Diodes sub. min . IN914 and IN9I6 type | 50p |
| B88 | 50 | Sil. Trans. NPN, PNP <br>  | 50p |
| $\overline{81}$ | 50 | Germanium Transitors | 50p |
| н6 | 40 | $\begin{aligned} & \text { 250mW. Zeneí Doiodes } \\ & \text { DO-7 Min. Glass Tyos } \end{aligned}$ | 50p |
| Н10 | 25 | Mixed volis, it wast Zeners Top hat type | 50p |
| मा\% | 20 | 3 amp. Silikon Stud Rectifiers. mixed volss | 50p |
| H13 | 30 | Top Hat Silicon Reccifiers, 750 mA . Mixed valts | 50 p |
| साठ | 8 | Experimenters" Pah of incesrated | Op |
| H20 | 20 | By126/7 Type Silicon Rectiflars I amp plascic. Mixed volts. | p |

FULLY TESTED AND MARKED SEMICONDUCTORS

|  | 80 | - | 5 |
| :---: | :---: | :---: | :---: |
| Ac107 | 0.15 | -ciro | 0.23 |
| ${ }^{\text {ACP }} 126$ | 0.15 | -c171 | 0.23 |
| ${ }^{\text {ACPI } 127}$ | 0.17 | OC200 | 0.25 |
| ${ }_{\text {A }} \mathbf{C} 128$ | 0.15 | OC201 | 0.25 |
| ${ }^{\text {ACl7 }} 1$ | 0.20 | 26301 | 0.13 |
| Acri7 | $0 \cdot 20$ | 26303 | 0.13 |
| A239 | 0.30 | 2N7II | 0.50 |
| AFIP6 | $0 \cdot 20$ | $2 \mathrm{~N}^{1102}$-3 | 0.15 |
| AFF39 | 0.30 | 2N1304-5 | 0.17 |
| BC154 | $0 \cdot 20$ | 2 N 130067 | 0.20 |
| BC107 | - 10 | ${ }^{2 N 1308-9}$ | 0.23 |
| ${ }^{8 \mathrm{BCClOg}}$ | $0 \cdot 10$ | 2NJB19FET | . 45 |
| ${ }_{\substack{8 C 109 \\ \text { BFI94 }}}$ | $\bigcirc$ | Powar |  |
| ${ }_{81274}$ | 0.20 | Tranziesors |  |
| beyso | 0.15 | OC20 | 0.50 |
| BSY25 | 0.13 | OC23 | ( $\begin{aligned} & 0.30 \\ & 0.35\end{aligned}$ |
| BSY26 | 0.13 | -0.26 | - 0.25 |
| ${ }^{\text {Bran }}$ | 0.13 | $\bigcirc$ | 0.30 |
| ${ }^{\text {BrY28 }}$ | $\bigcirc$ | OC35 | 0.25 |
| ${ }_{\text {BSYOSA }}$ | - 0.10 | OC36 | 0.37 0.30 |
| OCA1 | 0.13 | ${ }^{\text {A OUY } 10}$ | - |
| OC44 | - 0.13 | 25034 | 0.25 |
| Oct | - 0.10 | ${ }^{2} \mathbf{N 3 0 5 S}$ | 0.50 |
| $\mathrm{OC7}$ | 0.10 | Diodas |  |
| ося | 0.13 | MY42 | 0.10 |
| ${ }^{\circ} \mathrm{OC810}$ | 0.13 | OAM95 | 0.09 |
| ${ }^{\circ} \mathrm{CB3}$ | 0.18 | OA79 | 0.09 |
| OC139 0 0 | - | ON814 | - 0.00 |

## F.E.T. PRICE BREAKTHROUGH!!

This field effect transistor is the 2N3823 In a plastic encapsulation, coded as $3823 E$. te is also an excelData sheet supplied with device. $1-1030 \mathrm{p}$ each, $10-5025 \mathrm{p}$ each, $50+20 \mathrm{p}$ each.

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$$ <br> 24, 27, 30, 36, 43 Volts <br> 10 -watt Zener Diodes 5.1 . <br> $8.2,11,13,16,24,30$. <br> 100 Volts <br> $\begin{array}{lll}20 p & 17 p & 15 p \\ 25 p & 20 p & 15 p\end{array}$ <br> $\begin{array}{lllll}\text { Micro Switches, S/P, C/O } & 25 p & 20 p & 15 p \\ \text { f-amp Bridge Rec's } & 25 \text {-vale } & 25 p & 22 p & 20 p\end{array}$ <br> INTEGRATEO CIRCUITS <br> SL403D Audio Amp.3-Wat ts 2.00 $\quad 1.95 \quad 1.80$ 709 C Linear Opp. Amp. 25p 20p 15p Gates, Factory Marked and Tested by A.E.I. <br> J. K. Flip-Flops Factory. <br> Marked and Tested by <br> A.E.I. Decade Counser <br> L914 Dual 2 I/P Gate <br> $\begin{array}{lll}20 p & 18 p & 13 p \\ 50 p & 45 p & 40 p\end{array}$ $\begin{array}{lll}30 p & 45 p & 40 p \\ 40 p & 35 p & 30 p\end{array}$

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