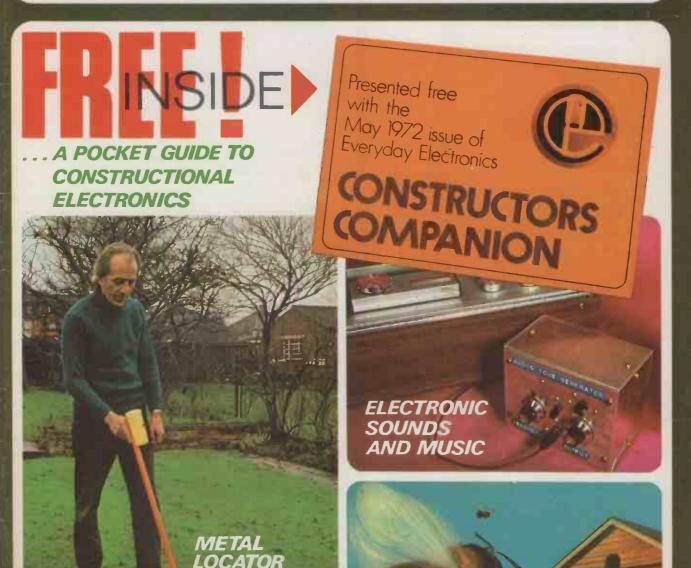
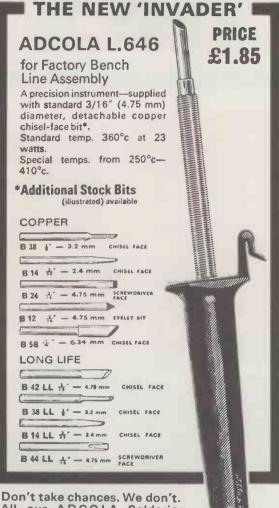
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STEREO AMPLIFIER Type SHV-2 x 3 watts

Fully built. Separate vol., bass and treble controls each channel; 12 × 4½ × 6in high, EZ80, ECC83, 2 × ECL86 valves. 0,P. trans. for 3-ohm speakers. Double wound mains trans. Suitable for crystal, magnet cartridge, tuner, etc. 200-250V. A.C. mains. 27, 50p P. & P.



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3 Wave band long-med.-short, Gram., 200-250V. A.C. Ferrite aerial. Chassis 13 × 7 × 5 in, Dial 13 × 4 in. Duble wound mains transformer 5 valves ECBA, EFS9, EBC31, ELS4, EZS0. Price \$10-63, (37p P. & P.) Output trans, for 3-ohm speaker.

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for only £52+£3.50 p&p Available complete

£22.00+90p p&p £32.00+£3 p&p £23-00+£1-50 00.773 SYSTEM 2
Viscount R101 amplifier
Viscount R101 amplifier
222
22 Duo Type III speakers
Garrard SP25 Mk. III with MAG.
cartridge, plinth and cover
£23

for only £69+£4 p&p Available complete Total

2 x Duo Type II speakers, pair £14·00+£2 p&p Garrard SP25 Mk. III with CER. and cover £52.00 SYSTEM 3 Viscount III Amplifier R100

for only £49+63.50 p&p Available complete

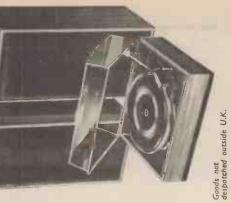
3 ohms. Simulated Teak cabinet. £14 pair +£2 p&p. £23 · 00 + £1 · 50 629.00

Size approx. 17" x 103" x 63". Drive unit 13" x 8" with parasitic tweeter. Max. power 10 wafts,

power 20 watts at 3 ohms. Freq. range 20Hz to 20kHz. Teak veneer cabinet. £32 pair+£3 p&p. Drive unit 134" x 84" with H.F. speaker. Max. Duo Type III. Size approx. 234" x 114" x 94"

SPECIFICATION R101

14 water per channel into 3 to 4 ohms. Total distortion @ 10W @ 1kHz 0-1%. P.U.1 (for ceramic cartridges) 150mV into 3 Meg. P.U.2 (for magnetic cartridges) 4 mV word kHz into 4 Ms. equilistes within ±148 R.I.A.A. Radio 150mV into 220K. (Sensitivities given at full power). Top out facilities: head-phone socket, power out 250mV per channel. Tone controls and filter characteristics. Bass: +1248 to — I/d8 @ 60Hz. Bass filter: 6d8 per octave cut.
Treble conrect retable + L2d8 to — L2d8 @ 18kHz.
Treble filter: 12d8 per octave. Signal to noise ratio.
(all conrects at mass 7 R101—P.U.1 and radio—65d8.
P.U.2 — 58d8. R100 same as R101 bur P.U.2 (for crystal carridges) 450mV into 3 Meg. Gross talk better than —35d8 on all inputs. Overload characteristics better than 26d8 on all inputs. Size approx 13½" x 9" x 3¾".



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AC127 AC128 AC176	17p	BFX29	38p
ACI28	18p	BFX84	25p
AC176	22p	BFX88	30p
ACI87	28p	BFY50	2lp
ACI88	27p	BFY51	21p
ACY19	23p	BFY52	22p
AD149	47p	MATIOO	25p
AD161/162	72p	MATIOI	29p
ADTI40	62p	MATI20	25p
AFI18	45p	MAT121	29p
AFI24	220	OC28 OC35 OC44	58p
AF125	19p	OC35	48p
AF126	20p	OC44	12p
AF127	19p	OC45	12p
AFI78	67p	OC71	HP
AF179	66p	OC72	12p
AFI80	45p	OC72 OC75	20p
AF239	32p	OC200	27p
BC107	Hp	OC201	38p
BCIOB	Hp	OCP71	60p
BC108 BC109 BC147	Hp	5T140	15p
BC147	120	ST141	23p
BC148	120	UT46	35p
BC149	12p	2N696	15p
BC157 BC158	15p	2N706A	120
BC158	14p	2N2926G	14p
BC159	14p	2N2926Y	13p
BD131	75p	2N29260	12p
BD132	75p	2N3053	25p
BFI15	25p	2N3054	60p
BF178	32p	2N3055	72p
BF179'	56p	2N3702	15p
BFI80	30p	2N3703	14p
BFISI	32p	2N3704	15p
BFI84	30p	2N3705	I4p
BF185	32p	2N3706	14p
BFI94	14p	2N3711	14p
BFI95	14p		
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DIODES

AALI9	Hp	OA202	10p
QA47	71P	BY100	15p
OA90	7 p	BY127	22 p
OA91	6p	BYZ12	22 P

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From 2 to 33 volts, 400mW, 15p; 1.5W, 22ip

SILICON BRIDGE RECTIFIERS

40 P.I.V., 1-5A 200 P.I.V., 2-0A

FUSES AND HOLDERS

liin glass—2;p
60, 100, 150, 250, 500, 750mA; 1,
125, 1-5; 2, 2-5, 3, 5, 7-5, 10, 15 amp.
fin glass—2;p
100, 250, 500mA; 1, 2-5 amp.
Anti-surge 1;in—5p
250, 500, 750, 850mA; 1, 1-5, 2, 3
amp.
Anti-surge 20mm—5p
Anti-surge 20mm—5p
80, 125, 200, 315, 400, 500, 630,
800mA; 1, 2 amp.

PANEL FUSEHOLDERS

For 1½in fuses 18

CONTROLS, Log. or Lin. Single, less switch, 15p Single, D.P. switch, 24p Tandem, less switch, 40p Skn., 10kn., 25kn., 50kn., 100kn., 250kn., 500kn., 1Mn., 24n

RESISTORS

Carbon All 5%, high-stability, E12 values. ‡W, tip; IW, 4p; 2W, 6p Wire-wound 5W, 10p; 10W, 12p

ELECTROLYTICS

1µF	450V	19p	1,000µF	25V	27 p
2µF	500V	20p	1,000µF	SOV	39p
4µF	350V	14p	2,000µF	25 V	36p
8µF	450 V	16p	2,000µF	50V	53p
16µF	450V	17p	2,500µF	25V	45p
25µF	50V	8p	2,500µF	SOV	60p
32µF	450V	24p	3,000µF	25V	48p
50µF	SOV	10p	5,000µF	25 V	55p
100µF	25V	10p	5,000µF	SOV	98p
100µF	SOV	10p	8-8µF	450V	18p
250µF	25 V	120	8-16µF	450V	20p
250µF	50 V	17p	16-16µF	450V	27 p
500µF	25V	18p	16-32µF	450 V	63p
500µF	SOV	25p	32-32µF	450V	49p
			50-50µF	350V	38p

MINIATURE ELECTROLYTICS

MILLI	AIURE	FLECT	ROLY	TICS
1 ₁₁ F	25V	10µF	64V	
2-54F	64V	16µF	40V	
4µF	40V	25µF	25V	
5µF	64V	30/4F	15V	47 (1)
BUF	ISV	50µF	15V	
BUF	40V	100µF	ISV	
10µF	15V			

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Type	Length	Width	Depth	Price
GB7°	2∄in	5±in	1±in	38p
GB8*	4in.	4in	-1-in	38p
GB9*	4in	2∄in	1 lin	38p
GB10*	4in	5±in	Lin	44p
GBII	4in	2+In	2in	38p
GB12	3in	2in	lin	33p
GB13	6in	4in	2in	52p
GB14	7in	- 5in	2∔in	63p
GB15	8in	6in	3in	Blp
GB16	10in	7in	3in	92p
	There sives for	seandard	Verahas	rede

VEROROARD

1511000	11/10	
Size	0-1 matrix	0-15 matrix
2∮în × 3∄in	22p	16p
2-in × 5in	24p	25p
3fin × 3fin	24p	25p
331n × 5in	27p	29p
17in × 21in	75p	57p
17in × 3∄in	£1	75p
Pinsbot	h sizes; packet	or 30, 18p

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PP75 Mains power supply, output £1.95
7½V d.c.
Both units are complete with cable and 5 pin D.I.N. plug

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Valve and Transistor Data book, 9th edition, 75p
Transistor equivalent book, BPI, 40p

LOW-OHM RESISTORS

 $2\frac{1}{2}$ watt wire-wound. 1Ω , 1.8Ω , 2.7Ω , 3.3Ω , 3.9Ω , 4.7Ω , 5.6Ω , 6.8Ω , 8.2Ω



CAPACITORS

					0.003/11	3004	Act.	29
	2-2pF	500V	S/M	7ip	0.0033µF	125V	P.S.	6p
	3-3pF	\$00V	S/M	71P	0.0033µF	S00V	Poly.	6p
	SpF	SOOV	5/M	71P	0.0033µF	1,000V	MDC	6р
		125V	0,5					
	10pF		P.S.	5p	0.0036µF	SOOV	S/M	15p
	10pF	500V	S/M	71p	0.0047µF	125V	P.S.	9p
	15pF	125V	P.S.	Sp.	0.00474F	500V	Poly.	6p
	15pF	500V	Cer.	4p	0.0047µF	500V	S/M	20p
	18pF	500V	S/M	710	0.0047		MDC	
				71P	0.0047µF	1,000V		6p
	22pF	125	P.S.	5p	0.005µF	100V	Mylar	3p
	22pF	500V	S/M	7 P	0.005µF	500V	Cer.	5p
	25pF	500 V	S/M	710	0.0068µF	125V	P.S.	10 p
	27pF	500V		135				127
			Cer.	4p	0.0068µF	500V	S/M	30p
	33pF	125V	P.5.	5p	0-0068µF	500V	Poly.	6р
	33pF	500V	S/M	71p	0.0082µF	125∨	P.S.	101p
	39pF	500V	S/M	7 P	0.0082µF	500V	S/M	30p
	47pF	125V	P.S.			127	Disc	· 4p
				5p	0.01 µF			
ı	47pF	500V	Cer.	4p	0.01 µF	125V	P.S.	10 lp
ı	50pF	500V	S/M	7 p	0.01 µF	160V	Poly.	4p
į	56pF	500 V	S/M	7 p	0-01 uF	250V	M.F.	3 p
ı	68pF	125V	P.S.	5p	0.01µF	400V	Poly.	3p
i				71-			Coly.	34
Į	68pF	500 V	S/M	71p	0.01µF	500V	Cer.	Sp.
	75pF	500 V	S/M	71P	0.01µF	500V	5/M	30p
i	82pF	S00V	S/M	71P	0.01µF	600V	MDC	7p
i	100pF	125V	P.S.	5p	0.01µF	1,000V	MDC	9p
١	100pF	500V	S/M	71P	O OIE.E	160V	Poly.	32
					0.015µF			3p
	100pF	500∨	Cer.	3p	0.015µF	400 V	Poly.	3p
	120pF	500V	S/M	7 P	0.02µF	100V	. Mylar	3p
ı	ISOpF	125V	P.S.	5p	0.022µF	18V	Disc	Sp.
	150pF	500V	5/M	71P	0.022µF	250V	M.F.	3p
			3/1-1	, ib	0.0224			3 P
	150pF	500 V	Cer.	5p	0.022µF	400 V	Poly.	3p
	180pF	500 V	S/M	7 P	0.022µF	600V	MDC	71p
	200pF	500 V	S/M	71p	0.022µF	1,000V	MDC	9n
	220pF	125V	P.S.	112	0.033µF	250V	M.F.	45
				\$p	0.033μ			4p
	220pF	S00V	Cer.	5p	0.033µF	400V	Poly.	4p
	250pF	500V	S/M	8p	0.047µF	12V	Disc	6p
	270pF	500V	Cer.	5p	0.047µF	160V	Poly.	3p
	300pF	500V	S/M	8p	0.0474F	250V	M.F.	3p
			3/11					
	330pF	125V	P.5.	5p	0-047µF	400V	Poly.	4p
	330pF	500 V	SIM	θр	0.047ggF	600V	MDC	8p
	390pF	500V	S/M	8p	0.047µF	1,000V	MDC	10p
	470pF	125V	P.S.	5p	0·1µF	30V	Disc	6р
							DISC	
	470pF	750V	Disc	5p	0·1μF	250V	M.F.	4p
	500pF	500V	S/M	8p	0-1µF	400V	Poly.	5p
	560pF	S00V	S/M	8p	0-14F	600V	MDC	10p
	680pF	125V	P.S.	6p	0·1µF	1,000V	MDC	13p
ı		500 V	S/M			250V	M.F.	130
ı	680pF			8p	0·15µF			5p
	820pF	500 V	S/M	8p	0-22µF	160V	Poly.	6р
	0.001µF	100V	Mylar	3p	0-22µF	250V	M.F.	5 p
	0.001 µF	125V	P.S.	6р	0-22µF	400V	Foil	10p
	0.0014F	400V	Poly.			1.000V	MDC	15p
				3р	0.22µF			130
	0.001 µF	500 V	S/M	10p	0-33µF	250V	M.F.	8p
	0.001 µF	500 V	Cer.	5p	0-47µF	250V	Foil	8p
ı	0.001 uF	1.000V	MDC	6р	0-474F	400V	Foil	15p
ı	0.0015µF	400V	Poly .	3p	0-47µF	1,000V	MDC	20p
ı								
	0.0015µF	500V	S/M	10p	1-0µF	250V	M.F.	15p
	0-0015µF	500V	Cer.	5p				
	0.0018µF	500 V	S/M	10p	Note:			
	0.002µF	100V	Mylar	3p	S/M - ei	lver mica	10/. 20	1
					D C - 31	- Lucation	- 210/	0.01
	0.002µF	V002	Cer.	5p	r.5. = p	olystyren	E 44 /0	tol.
	0-0022µF	125V	P.S.	6р	MDC-	a.c. ratin	g = 300	٧.
	0.0022µF	500 V	S/M	10p	M.F. = 1	Mullard m	in. foil	
	0.0022µF	1,000V	MDC	6p		eramic.		
	3.0022HP	1,000	1100	ob	Cer. = 0	c. aiiiic.		

PLUGS

LFAGO
Car aerial
Co-axial
D.I.N. 2 pin (speaker)
D.I.N. 3 pin
D.I.N. 4 pin
D.I.N. 5 pin, 180
D.I.N. 5 pin, 240°
D.I.N. 6 pin
Jack, 2+mm unscreened
Jack, Aymin discreened
Jack, 2+mm screened
Jack, 3+mm unscreened
Jack, 3 mm screened
Jack, kin unscreened
Jack, zin screened
Jack, stereo, unscreened
Jack, stereo, screened
Phono, plastic top
Phono, plated metal
Phono, fitted 4 ft lead
Wander, red or black
Banana 4mm, red or black

LINE SOCKETS

Car aerial
Co-axial
D.I.N. 2 pin (speaker)
D.I.N. 3 pin
D.I.N. 5 pin, 180
D.I.N. 5 pin, 240
Jack, 34mm
Jack, tin screened
Jack, stereo, screened
Phono, plated metal
I Hono, placed meras

SOCKETS

Tab		-
20p	Car aerial	Вp
35p	Co-axial, surface	8p
5p	Co-axial, flush	8p
12p	D.I.N. 2 pin (speaker)	10p
8p	D.I.N. 3 pin	9p
3p	D.I.N. 5 pin, 180°	9p
6p	D.I.N. 5 pin, 240°.	9 p
	Jack, 2+mm	10p
	Jack, 3+mm	10p
14p	Jack, tin unswitched	15p
	Jack, zin switched	17p
17p		
15p	Jack, stereo, switched	24p
16p	Phono, single	5p
16p	Phono, 2 on a strip	7p
16p	Phono, 3 on a strip	9p
15p	Phono, 4 on a strip	10p
43p	Wander, single, red or black	
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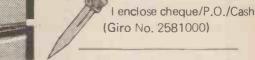
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Code	Power	Tolerance	Range	Values ovoilable	1 to 9	10 to 99	100 up
c	1/20W	5%	82 Ω-220K Ω	E12	9 (300)	8.0	7 7
č	1/8/	5 %	4.7 Ω-470Κ Ω	E24	í	0.8	0.7
č	1/400	10%	4·7 Ω-10M Ω	EI2	i	0.8	0.7
č	1/2//	5 %	4·7 Ω-10M Ω	E24	i - 2	ĭ	0.9
č	iw	10%	4.7 Ω-IOM Ω	E12	2.5	. 2	1 . 8
MO	1/2//	2%	ΙΟ Ω-ΙΜ Ω	E24	4	3 - 5	3
ww	IW	10 % ± 1/20 Ω	0 · 22 Ω – 3 · 9 Ω	EI2	7	7	6
ww	3 W	5%	12 Ω-10Κ Ω	E12	7	7	6
ww	7W	5%	12 Ω-10Κ Ω	E12	9	9	8
Codes	C - carb		hilim law sairs				

WW = wire wound, Plessey.

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E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades.

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everyday electronics

PROJECTS...
THEORY....

FOR ALL SEASONS

Our cover this month has quite an outdoor touch. Of course, you don't have to be an apiarist to sense that things are beginning to buzz in the outside world. Spring is now well advanced and thoughts are likely to be turning towards all kinds of pastimes and occupations for the coming summer months.

It is an appropriate time to point out that do-ityourself electronics has no closed season. Outdoor activities like gardening, touring, camping, sporting events, and so on, present many unique opportunities for putting electronics to effective use. So we advise, take stock now, anticipate your needs and start building to remedy any deficiencies in this respect.

GOOD COMPANION

The Constructors Companion given free with every copy of this month's EVERYDAY ELECTRONICS is small and compact. It has been designed for your pocket, so that wherever you go you can have essential facts constantly at hand. Compiled with the beginner particularly in mind, this booklet will prove a valuable aide-memoire for the more experienced constructor as well.

Those still feeling their feet will be glad of the technical back-up they can instantly call upon

when confronted with a choice of allegedly alternative or equivalent parts when shopping personally for components.

READY ACCESS

Our regular readers will already appreciate the amount of important and useful information they are accumulating, as the months go by. True, not everyone will have an immediate need for every project described. But a word of advice: do not discard back numbers. You never know when circumstances may arise that create a definite need which some previously described project would satisfy exactly.

This leads us on to another common problem: how to store numerous copies of a magazine so that ready access may be made at any time to one particular article. The only really satisfactory solution is to keep copies of the magazine in the binder specially designed to hold 12 issues of Everyday Electronics and which is now available.

Fred Bennett.

Our June issue will be published on Friday, May 19

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... EASY TO CONSTRUCT ... SIMPLY EXPLAINED



VOL. I NO. 7

MAY 1972

CONSTRUCTIONAL PROJECTS

AUDIO TONE GENERATOR Suitable for making electronic music by F. C. Judd	358
BEE COUNTER An electronic eye for the hive by G. A. Cozens	376
METAL LOCATOR Detects and locates metal objects underground by D. Bollen	382

GENERAL FEATURES

EDITORIAL	356
SHOP TALK New products and this month's constructionals by Mike Kenward	362
MAKING ELECTRONIC SOUNDS AND MUSIC Using the Audio Tone Generator by F. C. Judd	363
PLEASE TAKE NOTE	367
THEY MADE THEIR MARK No. 1—Introduction By J. E. Gregory	368
TEACH-IN Part 7—Semiconductors: Transistors by Mike Hughes	369
RUMINATIONS by Sensor	374
TEACH-IN HALF TERM TEST ANSWERS See how well you did	389
READERS LETTERS Your news and views 390	0, 393

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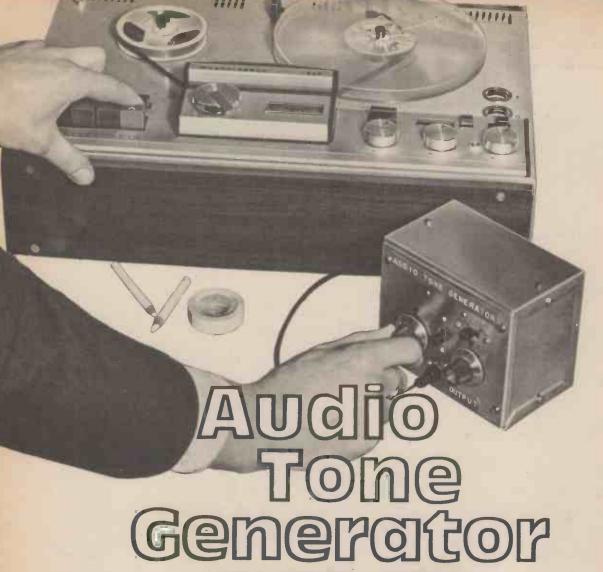
An Easi-Binder is now available handsomely finished in orange de luxe Balacron with black lettering on the spine. It holds 12 issues of Everyday Electronics.

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BY FRED JUDD

This simple tone generator covers the audio range from 50 to 2,000 Hz and has been specifically designed for use with a tape recorder to make electronic music.

The multivibrator is one of the most commonly used electronic oscillator circuits and generates an almost square waveform. It can be made to cover a wide frequency range without the need for switching in different values of components and moreover will produce a high output signal level for a relatively small supply voltage. As a primary signal generator it has many uses as a test instrument in audio as well as electronic applications.

The generator described in this article is used as a tone source for the creation of electronic music and "science fiction" sound effects in conjunction with a tape recorder. The feature Electronic Sounds and Music on page 363 deals with the use of the tone generator in detail.

GENERATOR CIRCUIT

The circuit diagram is given in Fig. 1 and employs three pnp transistors, two of which form the multivibrator (TR1 and TR2), the remaining one, TR3, being used as a squaring amplifier.

The operating frequency and mark to space ratio (see Fig. 2) of the multivibrator are set by the time taken for C2 and C3 to charge up enough to switch transistors TR2 and TR1 respectively. This "charging time" is determined by the value of the capacitor and the value of the resistor through which it is charged.

Providing the time taken for each capacitor to charge is similar then the mark to space

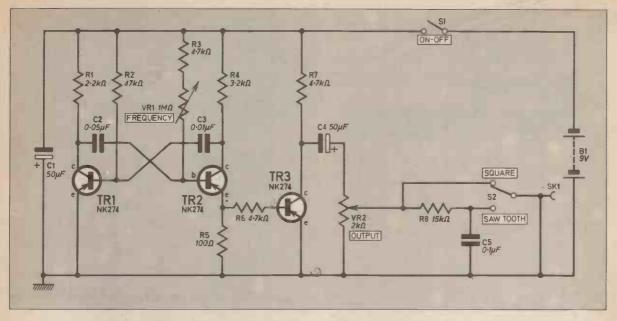


Fig. 1. Complete circuit diagram of the Audio Tone Generator.

ratio will be 1 to 1 or the mark and space will be of similar duration (Fig. 2). If we now change one of the controlling values—in this case VR1—both the frequency and mark to space ratio will be altered.

If we increase the value of VR1 the frequency will decrease as C2 will take longer to charge, and the mark to space ratio will alter for the same reason (see Fig. 3). Thus frequency control is achieved by VR2 and the total frequency range is approximately 50 to 2,000Hz.

The waveform has a mark to space ratio of 1 to 1 at approximately 1,500Hz at all lower frequencies the mark to space ratio increases becoming about 1 to 20 at the lowest frequency (Fig. 3).

The output from the multivibrator is taken from the emitter of TR2, through R6 to the base of TR3. Transistor TR3 is switched hard on and off by the output from TR2 and thus ensures a completely square output at its collector. The output level from TR3 is continuously variable from O to approximately 7 volts by VR2.

SAWTOOTH OUTPUT

The square wave output from TR3 can also be switched via S2 through an integrating network, C5 and R8, to provide an approximately sawtooth waveform (Fig. 4) of about 1 volt peak-to-peak maximum output, instead of the square-wave.

One of the major differences between a square wave and a sawtooth wave is the harmonic content and hence the tonal quality, when either are made audible via an amplifier and loudspeaker. The square wave contains only odd harmonics, in addition to its fundamental,

whereas a sawtooth wave consists of both odd and even harmonics plus the fundamental.

Audibly the square wave has a sound rather like that produced by a clarinet particularly in the region of middle C (261Hz approx.). The sawtooth wave has a sound rather more like a

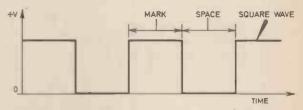


Fig. 2. A square wave with a 1 to 1 mark to space ratio.

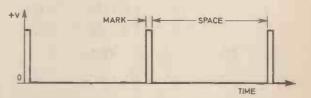


Fig. 3. A square wave with a 1 to 20 mark to space ratio.



Fig. 4. A sawtooth waveform.

Audio Tone Generator

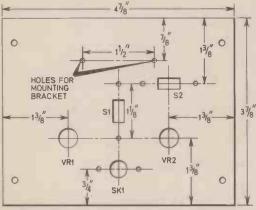
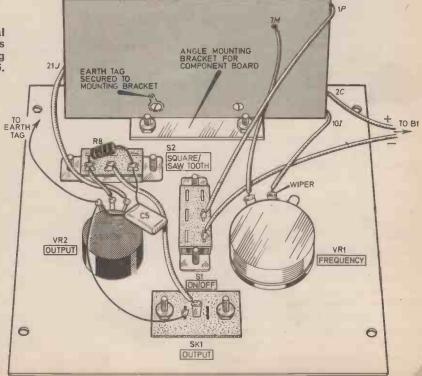


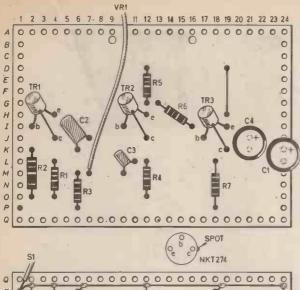
Fig. 5. Front panel drilling details.





Fig. 7. Wiring of the final unit. The tinted area is the component mounting board as shown in Fig. 6.





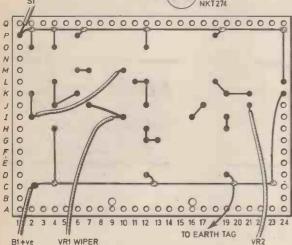


Fig. 6. Top and underside views of the component board. The transistor connections between the two diagrams are viewed from the underside.

flute. Both waveforms are used extensively in electronic organ voicing and for electronic music.

CONSTRUCTION

The prototype unit was housed in a box made from universal chassis parts. The pieces used assemble into a box measuring 5 by 4 by 3 inches. The sides and top and bottom can be assembled leaving one plate for the front panel and one for the rear. The plate used for the front panel is drilled as shown in Fig. 5 and is used to mount all the components.

If the layout and assembly of the generator is as shown there is just room in the case for a PP9 9 volt battery. Even if you spread the layout a little there should still be room for a slightly smaller 9 volt battery. The circuit board is 0.15 inch matrix plain perforated veroboard and is mounted on a 2 inch length of 3₈ by 3₈ inch aluminium angle.

Components....

Resistors

R1 $2 \cdot 2k\Omega$ R2 $47k\Omega$ R3 $4 \cdot 7k\Omega$ R4 $3 \cdot 2k\Omega$ R5 100Ω R6 $4 \cdot 7k\Omega$

4·7kΩ 3·2kΩ 100Ω 4·7kΩ 4·7kΩ

SEE

Capacitors

R7

R8

C1 50µF elect. 12V C2 0.05µF

All 4W ±10% carbon

C3 0.01 µF C4 50 µF elect. 12V

15kΩ

C5 0.1 / F

Transistors

TR1 NKT 274 germanium pnp TR2 NKT 274 germanium pnp TR3 NKT 274 germanium pnp

Potentiometers

VR1 1M Ω log carbon VR2 2k Ω lin carbon

Switches

S1 S.P.S.T. slide S2 S.P.D.T. slide

Miscellaneous

SK1 Phono socket B1 PP9 9V battery

Control knobs (2 off) Eagle type F10, case 5 x 4 x 3in made from universal chassis panels or a similar size case, battery connector, aluminium angle 2 x $\frac{3}{6}$ x $\frac{3}{6}$ in. Veroboard 5 x 4 x 0·15in matrix plain perforated, earth tag, connecting wire, 4BA fixings.

Commence wiring of the component board by inserting all components except the transistors, and the wire link on the top of the board as shown in Fig. 6. Turn the board over and connect up the two supply lines along the two sides of the board using 18 or 22 s.w.g. tinned copper wire. Next connect up the remaining components using the component leads where possible and connect the flying leads.

Finally insert the transistors checking carefully the lead connections with the underside view shown in Fig. 6, and solder them to the other components using a heat shunt on each lead as it is soldered. After checking the circuit board mount the board on the aluminium angle bracket and mount this on the front panel together with the remaining components.

Wire up all the components to the circuit board as shown in Fig. 7 and check the completed unit carefully before connecting the battery and switching on.

Continued on page 386



This month we have one item which many readers will probably wish to construct but which is not given in the form of a constructional project. It is the simple passive mixer that is described and drawn up in the Making Electronic Sounds and Music feature.

Since this is really a bonus that will be useful to those following the article we have not given full constructional details or a components list. All the component values are given on the circuit diagram and the wiring diagram shows how they are put together. The three sockets can be any type suitable for use with your particular tape recorder—the types we have shown are phono sockets.

The complete unit can be mounted in any small case. No battery or power supply is necessary. We would like to emphasise that this is a simple passive mixer and will not be able to cope with all inputs.

A more advanced type of mixer may form the subject of a future article. However this simple mixer should be suitable for use with the Audio Tone Generator that is also described in this issue.

Audio Tone Generator

There should be very few buying problems for the Audio Tone Generator. As described above the sockets could be changed to any suitable type if your equipment does not use phono sockets or if you already

have other types. Once again the case for this project can be any type that is available in a suitable size.

Bee Counter

We find it difficult to comment on the availability of cedar wood —not after-shave—but apparently this wood must be used or the bees will not accept it!

As far as the remaining components for the *Bee Counter* go make sure that the resistors you buy are of adequate wattage. The lamp and holder should be of the miniature type so that they can be accommodated in the wooden base panel. Since the current drawn by this circuit is fairly large the section in the article concerning the battery should be noted.

There are a number of Post Office type counters available so make sure you get the right one —4·2 ohms coil resistance is the important thing.

Metal Locator

The Metal Locator is a project which we are sure will create great interest but please remember that this is a simple one-transistor design and cannot be expected to out-perform a £30 unit. The use of Perspex or Paxolin is recommended for the locator head as these materials are not affected by damp or water.

All remaining components for the locator should be readily available. The use of a subminiature switch is recommended since only a small hole then needs to be cut in the plastic beaker. Any $50\mu A$ moving coil meter could be used in the locator provided it will fit the beaker lid. The one specified is probably the cheapest.

Finally do not forget the operating licence and don't say we did not tell you!

New Products

Two products from one goahead firm have been introduced this month. Both in the audio field, possibly the most outstanding is the Unisound 505 as Radio and T.V. Components call their do-it-yourself £25 stereo system. This competitively priced unit comes as a complete kit and only needs two screwdrivers to put together. All the electronics are in module form and are supplied with wiring looms that only need connecting up using a screwdriver supplied with the kit.

The large EMI speakers are housed in attractive cabinets again put together with only a screwdriver. It is said that anyone who can wire up a mains plug can put the system together in one evening. The system utilises modified Mullard Unilex modules, has an output of 3.7 watts continuous sine wave r.m.s. per channel; and frequency response of 40Hz to 20kHz at the 3dB down points. It would be very difficult to buy the individual components —including Garrard 2025TC deck. cartridge, plinth and cover and build a unit to match this one for £25, excluding the two speakers and cabinets.



The second unit from RT-VC is a £7 push button car radio kit, slightly more difficult to construct but any reader who has some experience of soldering should be able to build a working unit.

The kit is of good quality and uses the same push button tuning unit as radios costing three or four times the price. These features ensures good sensitivity and the pre-aligned i.f. (intermediate frequency) module and tuner avoid complicated alignment.

The kit is suitable for 12V positive or negative earth operation and readers may like to note that an after sales service—to repair any item not functioning correctly—is operated by RT-VC for all their kits; cost about £2 depending on the fault.





Simple experiments with a tape recorder

THE term "electronic music" almost defies explanation because it is not the music that is electronic but the equipment and methods of creating it.

Its origin goes back many years, in fact to the invention of the thermionic valve and even as early as 1921 a "concert" of electronic music was performed in Paris by an Italian, Luigi Russolo, who used what was then called electrical sound generating and reproducing equipment.

Electronic music was difficult to perform directly from sound generators, etc., because composition required arranging the sounds in a given order and even changing the order, and sometimes the sounds, at a later time.

MODERN METHODS

Magnetic tape recording finally provided the ideal medium for composition. The sounds required could be recorded and rearranged afterwards by simply cutting out the pieces of tape containing them and splicing these together again in the order required. This technique paved the way for composers who, with both electronics and magnetic tape at their disposal, could produce new kinds of music with tonal qualities never before possible.

More recently of course the music synthesizer has taken over the task of tone generation, etc., and electronic music composers can now programme a synthesizer, couple it to a tape recorder and produce "instant" electronic music.

Nevertheless there is much that can be accomplished by the amateur with an ordinary domestic tape recorder, an audio tone generator (like the one described on page 358) and some splicing tape. The techniques are simple and you can get a good deal of fun out of experimental electronic music and "science fiction" sounds.

Your efforts need not be wasted either because you can enter them for the experimental music and sounds section of the annual British Tape Recording Contest (details later).

EQUIPMENT

An ordinary spool to spool tape recorder is the main requirement and if you have a stereo recorder with provision for recording independently on either track or you can get together with a friend and use two tape recorders, so much the better. A tape recorder with track-to-track or duoplay facility is also advantageous especially if it permits echo effects.

It is not possible to lay down procedures for specific makes and types of tape recorder but you will find that most of the techniques described can be applied.

Note that cassette or cartridge tape recorders are of limited use for creative recording of this nature which requires fairly extensive tape

cutting and splicing.

Most modern spool to spool tape recorders are designed for stereo operation employing half or quarter track on standard quarter inch wide tape. If the tape recorder has a track-to-track recording facility it will have separate recording and replay heads, thus allowing a recording on one track to be copied on to another together with other signals.

Some stereo recorders may only have a common record/replay head which will not normally allow track-to-track copying but may have a facility for making separate recordings on each of two tracks. Information concerning such facilities should be given in the tape recorder instruction book. If in doubt, you should contact your dealer or the manufacturer for such information.

AUDIO TONE GENERATOR

An audio tone generator is not absolutely essential but is most advantageous. The simple Audio Tone Generator described on page 358 is quite suitable as it covers a wide enough frequency range and will deliver a square-wave or a nearly sawtooth-wave output signal, thus providing two basic sounds.

Sounds picked up by a microphone can also be used because these can be reshaped by tape cutting and splicing and by certain recording techniques. Magnetic tape will be required of course and for initial experimental work low

priced brands will suffice.

Some splicing tape and blank leader tape will also be required. Do not use ordinary plastic sticky tape, such as Sellotape, for splicing as

Fig. 1. (a) Original waveform of the recorded sound (b) The sound recorded and shown in (a) played in reverse.

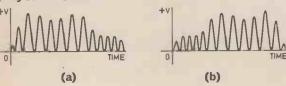




Fig. 2. Waveform of a sound which starts instantly and slowly dies away.

this may damage the tape and will not give a long lasting joint. Small kits of coloured leader and proper'splicing tape are readily available. A small tape splicer is also a very useful, though not essential, tool.

FIRST EXERCISES

It is important to know the extent to which your tape recorder can be used. If it has two or three speeds, as most of them do, record some musical sounds, whistling will do, at all three speeds and then play them back at one

speed only, say the highest.

The sounds recorded at the lower speed(s) will be raised in pitch, by one or two octaves, depending on the speed. If the replay speed is double that of the recording speed the pitch is raised one octave and the sounds will occur faster but if the replay speed is half the recording speed, the pitch will be reduced by one octave and the sounds will occur slower. This is one of the most simple but most used techniques.

REVERSE REPLAY

Now, if your recorder is a stereo machine try turning the tape over (reverse the spools) and see if you can obtain replay on another track in reverse, i.e., the sounds will be going backwards. This technique is also commonly used for electronic music because it alters the nature of the sound completely by placing what was the beginning of the sound, i.e., its attack, at the end as illustrated in Fig. 1 in which (a) is the sound as recorded and (b) as played in reverse.

If you cannot play sounds in reverse try this exercise; connect a tone generator, or if this is not available record whistles through a micro-

Photograph showing the use of a tape splicer to join up a number of sounds.





Recording various sounds, using the microphone, to form a composition.

phone. Start with the recording level control at the maximum, set the tape running to record the sound but then almost simultaneously slowly turn the record level control to zero.

On replay you will have a sound that starts instantly and then slowly dies away as in Fig. 2. With a little practice you will be able to get various dying away or decay times depending on the speed at which the recording level control is turned off. Now try the reverse procedure; gradually increase the sound whilst recording and then quickly stop it.

TAPE CUTTING EXERCISES

Now try some tape cutting; first use the highest tape speed and record a few sounds of different pitch, i.e., from a tone generator, or whistles via the microphone, each one lasting two or three seconds.

Locate the beginning of each sound on the tape by carefully feeding the tape across the head and then cut the tape about two inches in front of the sound. Run off the remainder until you reach the beginning of the next sound; cut the tape here and splice to the end of the piece containing the first sound. Cut and join pieces of the remainder of the sounds.

On replay you will have a series of short sounds each rapidly following the other. Now try a similar exercise but this time insert pieces of blank leader tape between each sound.

MUSIQUE CONCRÊTE

Finally a variation of the two previous exercises. Record a few sounds each at a different tape speed. These should preferably be musical sounds, such as whistles or tones, or sounds produced by tapping a wine glass for example. Cut one or two pieces of each from the tape and assemble them at random with pieces of blank leader between groups. The pieces may be long or short.

Try replaying the assembled tape at different speeds and note the effect. You are well on the way to a form of composition known as "musique concrête" which is the creation of abstract forms of music out of real sounds. The same technique can, however, be used for abstract forms of electronic music in which the main sound source is an audio tone generator.

USING A TONE GENERATOR

The exercises outlined above demonstrate how almost any recorded sound can be altered by tape cutting and by recording and replay at different tape speeds. Electronic music does not normally include natural sounds recorded via a microphone and therefore the sound sources are electronic, i.e., from tone generators of one kind or another. The recording and tape cutting techniques, however, remain the same.

If you have a full range audio signal generator then tones can be recorded at the pitch required. The simple generator described on page 358 has a frequency range of approximately 50 to 2000Hz.

If frequencies outside the range of the generator are required it is simply a case of recording and replaying at different tape speeds for example; if a frequency of around 4000Hz is required, record the highest pitch of the generator (approximately 2000Hz) at a tape speed of 334in/sec (inches per second) and replay at 712in/sec.

If a very low pulsing sound is required at say 20 to 25Hz record a square-wave signal from the generator at its lowest pitch and then replay the recording at half the speed. Some experiment in this direction will soon reveal the tonal and pitch ranges that can be obtained simply by recording and replaying at various tape speeds.

Once this has been done, further experiment with the audio tone generator can be carried out in order to discover the type of sounds that can be produced. Start by recording a continuous note and while recording this vary the frequency and output controls on the generator, try this for both the square and sawtooth outputs (note that the output in the sawtooth position is much lower than in the square wave position).

Try cutting and reversing the sounds recorded to obtain various effects. You can also try making recordings at a distorted level by turning up the record level control, this will distort the original sound and produce yet another effect. Try switching from one output waveform to the other whilst recording—you can vary frequency and output at the same time—and also try switching the generator on and off while recording, again you can vary the output and frequency whilst turning on and off.

Edit the sounds produced by cutting and splicing and experiment fully with all possible

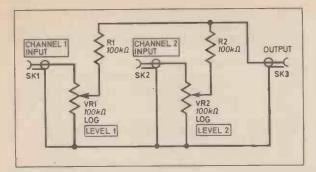


Fig. 3. Circuit diagram of a passive mixer that can be used for making electronic music.

effects. Once you have done this and feel fully conversant with the various effects that the generator is able to produce you can start to add one effect to another.

SIGNAL MIXING

Recording from track-to-track or using two separate tape recorders may necessitate mixing signals that are to be recorded and re-recorded i.e., signals from a recording already made to be mixed with signals from another source such as the tone generator.

Some recorders have built-in mixing facilities whilst others may permit a form of mixing by using the track-to-track recording facility or by superimposing one sound on another previously recorded. Again the tape recorder instruction book will provide information of this nature.

However, it is possible to build a very simple mixing circuit as shown in Fig. 3; Fig. 4 shows the construction. This is known as a passive mixing network, but will allow two signal sources to be mixed at different levels and coupled to a common input on a tape recorder (Fig. 5).

TAPE LOOPS

Another interesting technique widely used for electronic music is the tape loop. This is the use of a small endless loop of tape containing recordings which are played continuously to produce repeating rhythm patterns.

Record a few natural sounds, or low pitched tones from an audio generator, of quite short duration, one immediately after the other. Cut a piece of the tape containing the sounds, about 18 inches long, and splice the ends together so as to form a complete loop. Place the loop in the recorder so that it runs past the tape heads when the machine is set to replay. You can hold the loop under tension by one of the methods shown in the photographs. Try running the loop at different speeds and, if possible, reverse the direction.

Record some percussion sounds, e.g., sounds produced by knocking together empty boxes, etc. Cut out pieces and make up a loop consisting of the various sounds and blank leader tape.

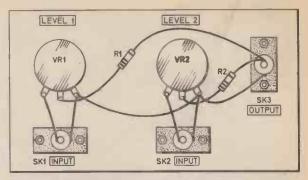


Fig. 4. Constructional details of the circuit shown in Fig. 3. Shop Talk refers to this figure.

For the first attempt use only two or three sounds and two or three pieces of leader.

You can make up an almost endless variety of fascinating rhythm patterns by this method and if you use two tape recorders the rhythm loop can be copied from one to the other whilst other sounds are added.

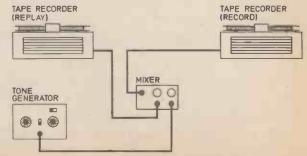
MULTIPLE RECORDING

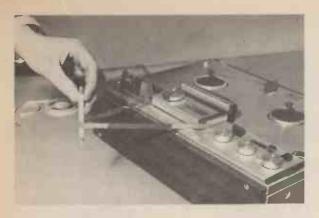
If you have a tape recorder with a track-to-track recording facility the scope is much wider as sounds may be recorded on one track and then re-recorded on to another track whilst adding more sounds. If your tape recorder can produce the echo effect this too can be used in various ways to produce those echoing science fiction sounds. Try allowing the echo to build up into a crashing roar and see if you can play it in reverse.

Now that you have discovered the variety of sounds and rhythms available using the facilities you have it is up to you to put these together to form an interesting "musical" passage. It may take some time before you achieve the required effect.

By combining even a few of the techniques outlined the number of permutations possible are fantastic. Instructions on composition cannot be given because no rules exist. Your ideas must come solely from imagination and experiment.

Fig. 5. Using the passive mixer to combine two signals for recording purposes.









Everyday Electronics, May 1972

The three photographs on the left illustrate various methods of using a tape loop. The top photograph shows a reversed loop held under tension by passing it around a pencil; this is only suitable for short periods.

The centre photograph shows a reversed loop held under tension by a small spool hanging over a table edge; this is only suitable for fairly large loops.

The lower photograph shows a system that can be used for any size loops by routing the tape around suitable objects—batteries are shown. This photograph also shows a cardboard tape holder used to keep recorded sections of tape in the order required.

COMPETITION

Finally, why not try an entry for the "technical experiment class" of the annual British Tape Recording Contest. It is open to anyone and the closing date for the 1972 contest is not until June 30. The Technical Experiment class allows for tapes of up to 4 minutes duration and includes; sound composition, electronic music, musique concrête, multi-track music and experimental sound recordings. The prizes are worthwhile and you can get an entry form free by writing to The Secretary, British Amateur Tape Recording Contest, 33 Fairlawnes, Maldon Road, Wallington, Surrey, and enclosing a stamped addressed envelope. You may also be interested to know that the special "Tape of the Year" award for 1971 was for an experimental class

Every tape entered is carefully assessed by the expert judges and their comments are passed to the contestant concerned when the tape is returned. Thus you will know how to make an even better tape next time.

PLEASE TAKE NOTE

The approximate cost of components given in the Simple Calculator article last month was incorrectly shown as £1.20. This should have been £2.20.

The probe flying lead in the Signal Injector article (March issue) should be soldered to Y3 not Y2 as stated in the text.

The Normatest 2,000 multi-range test meter mentioned in Shop Talk last month is available from: Croydon Precision Instrument Company, Hampton Road, Croydon, CR9 2RU.



THEY THEIR MARK

NO1 Introduction By J. E. Gregory

LECTRONICS is an internationally uniform world of symbols. Look at any advertisement or study the simplest circuit diagram in EVERYDAY ELECTRONICS and you will be confronted with strange symbols of every shape. Magical signs used to signify basic units of physical quantity; Table 1 lists some of them.

Although electronics is regarded as a modern science and hobby many of these units are named after pioneers, scattered throughout the world, whose accumulated research spans hundred.

dreds of years.

This series sets out to explain the symbol, and perhaps more important something of the man who gave his name to it. But let's begin our potted history of electronics at the beginning.

THE GREEKS HAD A WORD FOR IT

Take the word electronics itself, for that we must go back in time to ancient Greece. To the ladies of Greece passing time by decorating their spinning wheels with amber, found on shores in the far north. They observed that the amber when contacting the threads would draw the threads to itself as they separated from the wool, and then push them away in a frictional force. The

Greek word for amber was elecktron, from the verb elkein to attract. Although this phenomenon was observed and noted by several of the great Greek philosophers we have to jump two thousand years to the early 1600's and to the reign of Good Queen Bess, who was persuaded by her physician William Gilbert, to attend a demonstration of a frictional electric machine based upon the power of amber to attract. This power he called electricity.

was oon realised that the crackling and sparking of Gilbert's electric machine were the same phenomena on a minute scale, as thunder and lightning,

but how to prove it?

THE KITE FLYER

One of the first to try was the fifteenth child of an English immigrant; born in Boston Massachusetts in the year 1706, this was the well known American statesman and philosopher Benjamin Franklin (see illustration above).

His historic but dangerous

experiment trying to capture electricity from the sky occurred during a thunderstorm in the summer of 1752, when accompanied by his small son, he flew a kite with an iron door key. During the storm, he saw that sparks sprang from the key to his wrist, what he didn't realise of course was that if the lightning had actually struck the kite he would have been killed.

The study of natural phenomena had to take second place to his other activities, but he came to the conclusion that thunderstorms were simply the levelling of opposed electrical potentials, between one cloud and another or between a cloud and earth.

It was Franklin who introduced the positive and negative signs for electric charges, realising there are two kinds which neutralise each other.

Next month we move from America to 18th Century Italy and a scientist, Alessandro Volta, after whom the Volt, the measurement of electrical potential is named.

Photograph; Science Museum, London.

Table I FUNDAMENTAL UNITS

unit symbol	name of unit	physical quantity		
Α V F Ω W Hz	Ampere Volt Farad Ohm Watt Hertz	Electric Current Electric Potential Electric Capacitance Electric Resistance Power Frequency		
Н	Henry.	Inductance		

These basic units are often inconveniently large or small and the units are prefixed with the following symbols:

P	pico	÷ 1,000,000 million
n	nano	÷ 1,000 million
μ	micro	÷ I million
m	milli	÷ 1,000
k	kilo	× 1,000
M	mega	× I million
G	giga	× 1,000 million

Hence 5kV = 5,000 Volts; or 5mV = 0.005 Volt





TEACH-IN ... FOR BEGINNERS

By Mike Hughes M.A.



THIS year sees the twentieth birthday of the component most responsible for bringing electronics within the scope of do-it-yourself enthusiasts; it has greatly simplified design and construction and has also brought about terrific reductions in costs. It is the "transistor".

As a replacement for the valve, it allows us to use low voltages and removes the arduous task of having to assemble valve bases and massive transformers on tank like chassis. Connections to a transistor are few and the basic way it operates in a circuit is quite easy to understand.

PNP-NPN

The transistor is a member of the semiconductor family and is basically a sandwich of different types of either silicon or germanium. The "filling" of the sandwich can either be p- or n-type material; we can clad a p- type filling with n-type material giving what we call an npn transistor. Alternatively a pnp device is made by filling a p- type material with an n-type.

One encounters both types in practice but nowadays npn devices made from silicon predominate, the reason being that they are easier to make and hence cheaper!

Fig. 1(a) shows a diagramatic cross-section of both types of transistor, pnp and npn. One end is heavily doped and is called the "emitter"; the other end is lightly doped and called the "collector".

The filling material is very thin in practice (usually one or two microns; 1 micron is a

millionth of a metre) and is called the "base". In its simplest form you can think of an *npn* device as two diodes connected together by their anodes (back-to-back), and facing each other in a *pnp* device, Fig. 1(b).

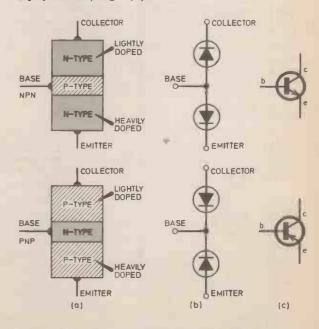


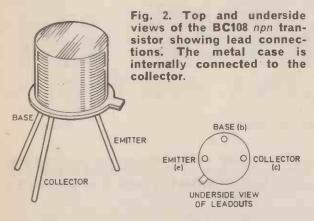
Fig. 1. (a) Schematic diagram of the internal make-up (b) equivalent representation and (c) circuit symbol for (top) *npn* transistor and (bottom) *pnp* transistor.

BASE CONNECTIONS

All the transistors you will come across have connections brought out from the emitter, base and collector. A very common silicon npn device is the BC108 and we shall be referring to this frequently in this series.

Fig. 2 shows what it looks like. If you have one handy see if you can identify which lead is

which.



The emitter is the one closest to the spigot on the side of the can, the collector is diametrically opposite, and the base is between the two but set off to one side. This is a metal can transistor and the can is electrically "live"—in actual fact it is connected to the collector as well as the lead out wire.

Different types of transistor may have different shaped cans and some are in plastic encapsulations. Always make sure you know which lead is which before you start using a transistor.

Most constructional projects in EVERYDAY ELECTRONICS give you lead designations for the transistors specified, but if you want to experiment with alternative types make sure you know the correct base lead connections.

SIMPLE TEST

Use the BC108 npn transistor to identify the effect of the two diodes connected back-to-back. First of all make an ohmmeter on the Demo Deck. Use a 4.5V battery (not 9V) in series with a 2.2 kilohm resistor and VR2 (5 kilohm). Complete the circuit and set VR2 to give zero ohms at full scale deflection and then connect the leads of your ohmmeter between the base and emitter connections of the transistor—to do this it is best to solder the transistor on to three adjacent pins of the Demo Deck and use crocodile clips on the leads from the meter.

If you connect the meter so that the lead coming directly from the negative terminal of the battery goes to the emitter, the meter needle will move to almost full scale showing there is little resistance in the transistor. Now reverse the leads so that the base is more negative than the emitter—you should see that no current

flows (indicated by meter needle not moving). Thus the base-emitter junction is a diode and follows the same rule that we saw last month.

Now leave the lead on the base and transfer the one from the emitter to the collector—again no current flows but reverse the leads and current flows between the base and collector.

If you connect the leads between the collector and the emitter no current should flow whichever way you have them because in both connections, the current would have to pass through a reverse biased diode.

This simple experiment can be used as a rough and ready test to check if a transistor is likely to be in working order, and provided you remember the rule "make p stand for positive for current to flow" you can use it to identify npn and pnp transistors.

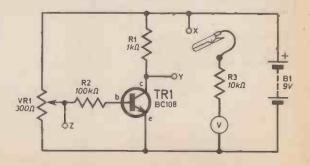
REVERSE VOLTAGE LIMITS

Like all diodes, the junctions of a transistor have reverse voltage limits. These are usually specified with abbreviations. For the BC108 the reverse emitter-base voltage ($V_{\rm ebo}$) is 5V—i.e. you must never make the base more than 5 volts negative with respect to the emitter (this is why we had to use 4.5V for our ohmmeter instead of the 9V we have been used to). Likewise the reverse base/collector voltage ($V_{\rm cbo}$) is 30V. You might expect the reverse voltage between the emitter and the collector to be equal to the highest of the other two but this is not the case—it is lower—for the BC108 $V_{\rm ceo}$ is 20V.

The "O" in the suffixes of the reverse voltage characteristics indicates that the third terminal is "open circuit" i.e. not connected.

HOW THE TRANSISTOR WORKS

Let's see what a transistor actually does by using the circuit of Fig. 3(a). Now that we are using the transistor in a real circuit it is important to note the polarity of the supply voltage—for an npn transistor the collector must always be kept more positive than the emitter (the converse applies to pnp devices). We are going to make the transistor work like a tap and control the amount of current flowing through R1. You can see this happening if you follow the details through on the Demo Deck.



VR1 is a 300 ohm potentiometer working as a potential divider giving us a variable supply at its wiper.

Wire up the circuit of Fig. 3(a) on the Demo Deck as shown in Fig. 3(b), but do not connect R2 to the base of the transistor just yet.

Resistor R3 and the lmA meter makes a 10V range voltmeter in the usual way. Connect the negative lead to the emitter of the transistor. All voltages we measure will be relative to that of the emitter.

First measure the power rail at point X—it should, of course, be +9V; now measure the potential at the collector of the transistor (point Y) it should be +8·2V. This is what is expected because no current can flow through the back-to-back diodes of the transistor, but the meter will draw some! If you had a high sensitivity meter (say 20 kilohm per volt) this current would be negligible and you would see +9V at both points, X and Y.

Now set VR1 so that the potential on its wiper is zero (with respect to the emitter) and connect R2 to the base of the transistor. VR1 potential is measured by attaching the crocodile clip from the meter to point Z. Again measure the potential at the collector—it should not have changed.

We shall now see what happens if we increase the potential at the wiper of VR1. Do this in 0.5V increments (use crocodile clip at point Z) and for each setting measure the collector potential. You should see that once the potential of the wiper exceeds 600mV, the potential at the collector falls, and continues to fall towards zero as the controlling voltage is increased. Once the collector potential reaches almost zero no more

control can be effected. We say that the transistor is now fully conducting between collector and emitter. This state is called "saturation."

Record your results and plot a graph of collector voltage versus voltage at the wiper. A graph should be obtained similar to that of Fig. 4.

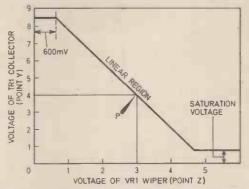


Fig. 4. The graph obtained by plotting the recorded results of experiment using circuit of Fig. 3(a), i.e. voltage at point Y versus voltage at point Z.

Control of the collector/emitter current is brought about by passing a current through the forward biased base/emitter junction. The more current we pass into the base in this way, the more current we can control between the collector and the emitter. The controlling current is called "base current," (I_b) and the controlled current "collector current," (I_c) .

Base current is set by the potential difference between the wiper of VR1 and the emitter of the transistor, acting through the resistance R2

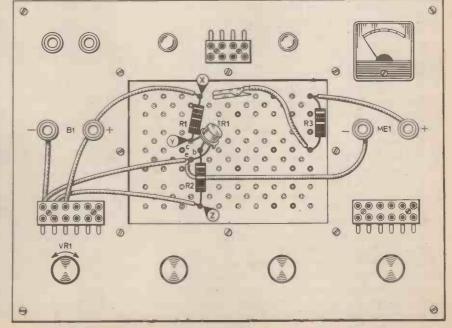


Fig. 3(a) (left). The circuit diagram used for investigating some of the properties of a BC108 transistor.

Fig. 3(b) right. The circuit of Fig. 3(a) wired up on the Demo Deck.

and any internal resistance between base and emitter. The latter is small and can be neglected at this stage. We must, remember, however, that the base must be made at least 600mV positive with respect to the emitter before any current can flow (this is the usual forward voltage drop for any silicon junction).

We can thus calculate the current flowing into the base by measuring the potential at the wiper of VR1, subtracting the base emitter forward voltage drop (600mV) and dividing by the value

of R2.

GAIN

If you do this for your experiment you will find that the base current ranges from 0 to 0.084mA. The range of collector current we are controlling was from 0 to 9mA. It can be seen that the transistor enables us to use a very small current to control a larger one. We call this effect "current amplification." The factor that governs the ratio between lb and lc is called "gain" and although it increases with Ic it is pretty well constant for any given transistor. It can, however, vary widely between different types of transistor and even between devices having the same type number! Provided you take a combination of base and collector currents within the controllable region (this is called "linear region") you can calculate the gain of the transistor you are using.

It would be best to increase the potential at VR1 until the collector potential is approximately 4V. This reduces the shunting effect of our voltmeter.

Use the precise values of voltage measured to calculate the current through R2 and R1 then use the ratio of these values to calculate the gain.

gain = collector current + base current

 $=I_{o}+I_{b}$

For the BC108 transistor it should be approximately 200, but as we have said, will vary from device to device.

Example To calculate the gain from your plotted curve (similar to the one of Fig. 4) select a convenient point on the linear region such as point P of Fig. 4.

The base current, I_b is given by the voltage difference between the base and emitter divided by the base resistor.

i.e.
$$\frac{3-0.6}{100.000} = 0.024$$
mA

Now the voltage drop across the collector resistor R1 is (9-4)V=5V. Therefore, collector current I_0 is (5+1000)=5mA.

Substituting these values for I_0 and I_b in equation (1) gives the gain $= (5 \div 0.024) = 208$.

There are various ways of describing current gain for a transistor so we shall define that measured above a little more precisely—it was the d.c. current gain. This is sometimes abbre-

viated to the designations β (beta) or $h_{\rm FE}$. The latter is rather a strange type of designation but is one of a range of what are called "h" parameters—we need not worry ourselves about these in this series except for the term $h_{\rm FE}$ which is usually used in manufacturer's data sheets. Do not confuse $h_{\rm FE}$ with $h_{\rm fe}$, the latter is called the small signal current gain and we shall not be dealing with this until later.

The gain equation above can be rewritten:

 $I_{\rm c} = h_{\rm FE} \times I_{\rm b}$

Remember that the experiment we have just done has been using a silicon npn device. We could have used one made from germanium having npn structure and obtained a similar effect—except that the base/emitter forward voltage drop would have been only about 200mV and $h_{\rm FE}$, in general, would have been lower.

We could also have used a silicon or germanium pnp device but would have had to reverse the battery connections so that the collector was negative with respect to the emitter. The same rules would have applied and we could have still calculated a value for $h_{\rm FE}$.

If you are a little confused by the difference between npn and pnp devices do not worry too much as this stage—most of the early experiments in Teach-In will use npn devices and when you have got used to these you will find it quite straightforward to switch over to pnp devices when necessary. The most important thing to remember is the polarity of battery voltage when using one type or the other. An aid to remembering what the polarity ought to be is to bear in mind the direction of conventional current flow;

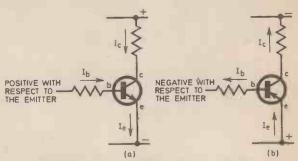


Fig. 5. Circuits showing major current flow directions for (a) npn and (b) pnp transistor. l_b —base current, l_c —collector current, l_e —emitter current.

the arrow on the emitter of the symbol points in the direction of current flow, i.e. it points away from positive and towards negative. See Fig. 5.

Whether using npn or pnp devices an aid to remembering how to turn collector/emitter current "on", is to make the potential at the end of the resistor connected to the base tend towards the same polarity voltage as applied to the collector; the more you move towards this voltage, the more I_b increases, and I_c will increase in direct proportion.

When the potential feeding the base rises towards the supply voltage the voltage at the collector falls towards the emitter voltage. This is called "inversion."

In Fig. 3 R1 is called the "collector, load." The limit of I_c control is set by the value of this resistor; if it has a high value then it does not matter how much base current you apply, you cannot control more collector current than that given by the collector supply voltage divided by the value of collector load. On the other hand, if the load is too low you might find yourself trying to force more collector current than the construction of the transistor can handle. Thus one of the specifications of a transistor is the maximum collector current it can handle without "blowing". This is called I_{cmax} and for the BC108 is 100 mA.

A final parameter we must deal with is the power rating of a transistor. As current is passing through it a certain amount of heat is dissipated. We already know that too much heat can spoil the properties of a semiconductor so it must be limited. The limit is set by the maximum power dissipation parameter, P_{cmax} . It is easy to calculate what the power dissipation is likely to be; it is the dissipation you would get if you replaced the transistor in the circuit with a resistor having the same ohmic value as the collector load.

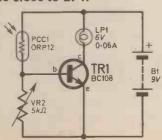
Table 1 gives you some typical values of parameters for some common transistors of varying types, powers and polarities.

Table 1: THE MORE IMPORTANT CHARACTERISTICS OF SOME COMMON TRANSISTORS

Type P	olari	ty Pc max	V_{cbo}	Vceo	$V_{\rm ebo}$	I _{c max}	hee
BC108	nþn	300mW	30V	20V	5V	100mA	240
2N2926	nþn	200 m W	187	18V	5V	100mA	150
BFY51	nþn	800mW	60V	60V	6V	IA	70
BFX13	pnp	300mW	-20V	-15V	-5V	100mA	120
2N3702	pnp	360mW	-40V	-25V	_5V	200mA	60
ACI26	pnp	500mW	-32V	-32V	-10V	100mA	100
OC72	pnp	125mW	I6V	-16V	_3V	125mA	50
OC26	pnp	12W	-16V	-16V	-10V	3-5A	50
OC36	pnp	30W	-80V	-32V	-40V	IOA	70

Fig. 6(a) (below). The circuit diagram of the "Electronic Candle" which illustrates positive feedback.

Fig. 6(b) (right). The circuit of Fig. 6(a) wired up on the Demo Deck. Ensure that PCC1 is close to LP1.



ELECTRONIC CANDLE EXPERIMENT

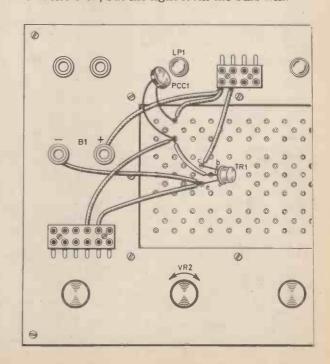
We shall now make a simple working circuit using the circuit diagram of Fig. 6(a). This is wired up on the Demo Deck as shown in Fig. 6(b). Connect the ORP12 (light dependent resistor) very close to the LP1 on the Demo Deck as shown below. Set VR2 to zero ohms. The potential at the base of TR1 will be zero, therefore no current will flow between collector and emitter. Now, in a reasonably lit room, increase the value of VR2. At a certain point the potential at the base will reach 0.6V (set by the potential dividing effect of PCC1 and VR2) and the transistor will start to conduct (the bulb will glow dimly).

Continue to increase the resistance of VR2; the current flowing through PCC1 will now pass into the base/emitter circuit of the transistor in preference to the higher resistance path through VR2. This base current will cause TR1 to pass more collector current until the bulb is fully illuminated.

When you reach this point (the minimum value of VR2 that will give full illumination) try casting a shadow over PCC1, the lamp will go dim and ultimately go out altogether as I_b reduces due to the resistance of PCC1. We did a similar sort of thing in Teach-In Part 4.

The difference is that we now have a circuit that is much more sensitive to small changes in light level which is brought about by the transistor amplifying the current from the photo resistive cell.

If you place the cell very close to the bulb in a dimly lit room you can set the value of VR2 so that the ambient lighting does not turn the transistor on, but the light from the bulb will.



Break the light path between the bulb and the cell and the bulb goes out and stays out. Now use a match or lighter to provide a stimulus of light. Bring it close to the bulb/cell assembly and the bulb lights up; you can now remove the match and the bulb will stay on because its own light output is holding the transistor on. This is called "positive feedback" and in this circuit will provide an amusing party trick—especially if assembled to look like a candle.

A bit of practice at "snuffing" the candle with the fingers (actually you are breaking the light path between the bulb and the cell) will make the effect even more astounding.

Photograph of the Demo Deck set up for the Electronic Candle Experiment showing the lamp being "lit" by the light emitted from the lighter.



TEACH-IN PART 6-ERRATA

Fig. 4(b) last month shows a lead connected wrongly. The lead from the junction of R3 and the negative meter terminal should go to the negative end of VR1 (not the wiper as shown) i.e. the one connected to the battery negative.



Next month: Multivibrators. The components needed for next month in addition to those already acquired are: resistors 22 kilohm (2 off), 100 ohm (1 off); capacitors 0·1μF polyester (2 off), 500μF elect. 12V (1 off); transistors BC108 (1 off); diodes QA91 (1 off).

Ruminations By Sensor

Not so Clever

The coal miners' strike has shown how dependent we are, in this age of high technology, on the efforts of men who work in damp, dirty and often dangerous conditions.

I find it difficult to comprehend that on one hand the semiconductor industry owes its existence to the ability to obtain
and to process materials with
an impurity content of less than
ten parts in a thousand million,
and to operate with tolerances
down to one millionth of a metre,
while on the other hand men
have still to dig fossil trees out
of the earth (albeit with mechanical assistance) so that these fossilised remains can be burnt to
boil water in order to raise steam

and to generate electricity! Without coal and electricity there would be no semiconductor industry; truly our idol has feet of clay!

Let There be Light

Have you heard about the old lady who telephoned the C.E.G.B. to complain that, during the power cut, the buses were passing her house with all their lights on? She also said that she could manage to get along quite well without the electricity, except for the little light in the hall, and could they please leave that one switched on.

Many people must have been irritated, in the early days of the strike, to see street lights blazing all day and switched off at night, due to their electric clock switch mechanisms getting umpteen hours behind. To the electronics man the answer to this problem is so simple—a light operated switch, either using discrete components or in integrated form.

A recently introduced inte-

grated circuit provides the necessary photo cell, level sensor and time delay all on one tiny chip of silicon and complete with lens. It could operate a relay or, better still, work into a switching transistor controlling the street lamp directly.

Some years ago, I was shown around a large generating station, where, tucked away in a dusty corner there was a cast iron box about the size of a domestic cooker. This apparatus was installed at the station about twenty five years ago and its purpose was to switch on all the electric street lamps in the town.

When switched on it produced a ripple which was superimposed on the mains. Sections of street lighting were grouped together under the control of master switches, spread throughout the town, which were operated by switching on the ripple equipment. The system had been in use but for some reason, unknown to my guide, had been discontinued. It would have been a blessing during February 1972.

Everyday Electronics, May 1972



On sale Friday, May 19



MODERN research calls for accurate measurement and comparisons, and with this in mind this device was designed to help the beekeeper assess the performance of his beehives more definitely, and to compare the different strains of bees under the same working conditions and so help to breed a strain which will produce the most honey under all the difficulties encountered in our changing climate without the rather nasty habit of the English bee, of attacking the bee-keeper as soon as he appears anywhere near the hive.



The Bee Counter is an instrument which records the number of bees entering the hive, and used in conjunction with other devices such as a wind speed indicator, a wind direction indicator, an air temperature thermometer, a maximum/minimum thermometer, a rain gauge and a sunshine recorder, then some degree of assessment can be made, and some basis established for the bee-breeder to work upon his main goal—lots of honey from a reasonably good tempered, busier bee.

The Bee Counter makes use of the fact that bees are highly organised in their habits, and utilises the bees sense of sight and smell. These bee "characteristics" are used in the design of the cabinet housing all the circuitry which is described later in full detail.

THE CIRCUIT

The complete circuit diagram of the counter is shown in Fig. 1 and is basically an amplifier which works as follows.

The lamp LPI, which is always "alight" when

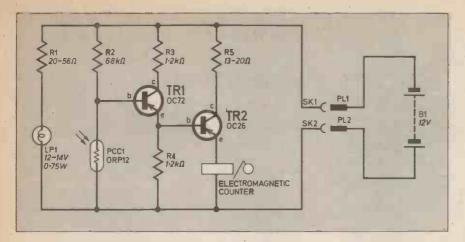


Fig. 1. The complete circuit circuit diagram of the Bee Counter.

the unit is switched on, illuminates the light dependent resistor, PCC1, and causes its resistance to be at a low value, about 100 ohms.

The l.d.r. and R2 form a potential divider circuit and under "illuminated conditions" of the l.d.r., a positive voltage with respect to the emitter, is applied to the base of TR1 causing it to be in a conducting state.

With TR1 conducting, a negative voltage is applied to TR2 base with respect to the emitter and consequently TR2 is "off" (not conducting).

When the light path between LP1 and PCC1 is broken, the resistance of PCC1 increases considerably (to about 100 kilohm for complete "blackout"). This causes the potential at TR1 base to go negative and turns it "off". This state of TR1 causes the voltage applied to the base of TR2 to go more positive and causes it to switch "on" i.e. conduct—current flows through TR2.

When current flows through the emitter leg of TR2 containing the relay coil in the counter,

the relay is energised.

When the light to PCC1 is restored, TR2 switches "off" and the counter is de-energised and springs back to its off position and in doing so mechanically adds "one" to the counter readout.

The arrangement of LP1 and PCC1 in the case is so devised that the bee, on entering the hive, breaks the light path between these devices

and its entry is thus recorded.

The 13-20 ohm 3 watt resistor, R5, in the collector circuit of the power transistor, TR2, is to prevent damage to the counter or the transistor if the entrance passage to the hive should become blocked, as once happened in the prototype when a drone got stuck in the narrow part.

A heavy duty battery is required to operate the Bee Counter since current drain is substantial -250 mA when TR2 is "off" and 400 mA when TR2 is "on" at 12V. A car battery is therefore recommended to supply the power. The cost of this battery is not included in approximate cost.

The voltage is fairly critical as it must be sufficient to operate the counter, but not high Components.

Resistors

R1 20-5612 3 watt

R2 68kΩ

R3 1 ⋅ 2kΩ

R4 1.2k()

R5 13-20Ω 3 watt

All ½ watt carbon - 10% unless otherwise

stated

Transistors

TR1 OC72 (or similar) germanium pnp

TR2 OC26 germanium pnp

Light Dependent Resistor PCCI ORP12

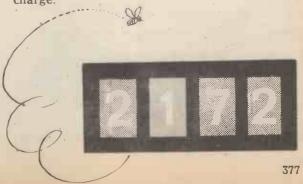
Micellaneous

LP1 12-14V 0.75W bulb and holder PL1, PL2 Wander plugs, 1 red 1 black (2 off) SK1, SK2 Sockets to suit plugs PL1, PL2 B1 12V battery—heavy duty rechargeable type

(Not accounted for in cost box.) Counter: Post Office type 14C 4·2Ω 4 figure readout. Cedar wood, Perspex and adhesive, Paxolin, wood screws, 4 B.A. nut and bolt, wood glue.

enough to cause overheating of TR2 or the counter coil in the event of the passage being blocked for long.

If the apparatus is disconnected every night the battery will last at least a week on one charge.



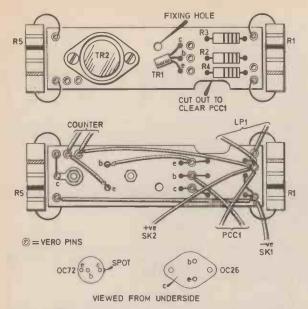


Fig. 2. The layout of the components on both sides of the Paxolin board. Veropins are used for attachment.

Variations in performance can be dealt with in several ways. The lamp should be bright enough to turn off the amplifier, but not any brighter than necessary. This is best adjusted by altering the series resistor R1, which may be increased to as high as 56 ohms.

Also, the size of the light hole can be varied, or a part of the l.d.r. painted over so that it has less area exposed, until the instrument is sufficiently sensitive, but positive in its action.

THE COUNTER

The electromagnetic counter used is a Post Office type. It has a four digit readout and can thus count up to 9,999. The maximum count rate is ten per second.

COMPONENT WIRING

Most of the components of Fig. 1 are mounted on a piece of Paxolin size $4^{1}_{2} \times 1^{1}_{4}$ inches with a cut-out as shown along one side to accommodate the light dependent resistor, PCC1.

Both sides of the board containing the components are shown in Fig. 2.



Veropins are used for mounting the components in position and small holes should be drilled where indicated to accommodate these pin's.

Three more small holes of the same size should be drilled to take the leads of TR1 as shown.

Drill the component board fixing hole and the four holes for transistor TR2; (see reverse side of component board Fig. 2); ¹8 in. diameter holes will do for all five holes.

Begin assembly by pushing in all the Veropins and then attach TR2 to the board using two small nuts and bolts.

The connection to the collector of TR2 is via its casing, so a solder tag should be attached to one of the securing bolts to enable this connection.

Attach and solder all the components, link wires and flying leads as detailed in Fig. 2 making sure a heat shunt is used when soldering in TR1, which incidently should be the last component connected.

The l.d.r. should be attached to the board via 6in. long flexible leads.

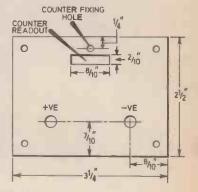
The flying leads to the counter should be about 4in. long.

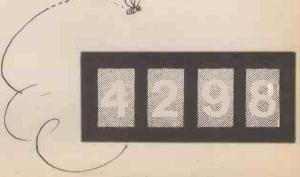
The two wander sockets used for battery connection to the counter, are attached to the end of the case which is made from a piece of Paxolin, dimensions are given in Fig. 3.

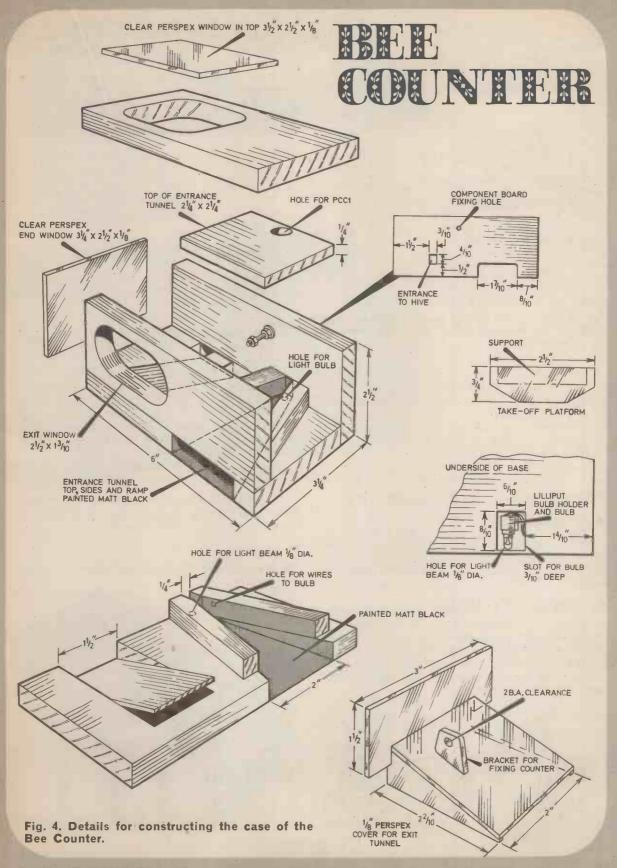
The connection wires from the wander sockets to the component board should be about 4in. long.

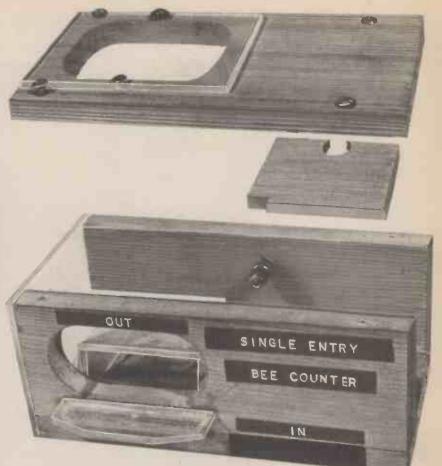
Connection to the battery is made via two wander plugs and a length of twin flex.

Fig. 3. Dimensions of the end made from Paxolin to accommodate the wander sockets for battery connection, and counterreadout.









A photograph of the prototype with top and tunnel lid (which holds PCC1 in position) removed. The photograph clearly shows the entrance and exit tunnels (labelled IN and OUT respectively). The take-off platform, made from Perspex, is located just beneath the exit cut-out, and is glued in position with Perspex adhesive.

EXIT AND ENTRANCE GEOMETRY

As said before, this device and its design utilises the bees' senses of smell and sight. From inside the hive, the exit from the hive appears as a bright opening to the outside world and so the exit path through the instrument must be a tunnel with transparent sides and top to allow this condition to be fulfilled.

In the instrument this tunnel slopes upwards so that when the bee emerges, it finds itself on a platform of Perspex, about ³4in. wide, situated above the hive base, and flies away.

When it returns, it will land on the hive base (landing/alighting board) and walk towards the hive.

The entrance to the hive is now through the Bee Counter which is a tunnel painted matt black; when the bee walks along the front of the instrument and reaches this tunnel it will enter.

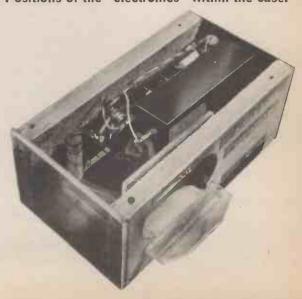
On entering, the tunnel becomes narrower and at the same time slopes upwards until it is just wide enough for a single bee to pass.

There is a lamp under the narrow part, with a hole in the floor of the tunnel, made up to the level of the floor with Perspex cement so that light can shine up through it.

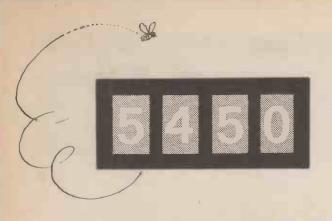
The light dependent resistor is situated in the

roof of the tunnel and as the bee walks between this and the lamp, the light beam is cut and the circuit activated.

Positions of the "electronics" within the case.



Everyday Electronics, May 1972



CONSTRUCTION OF CASE

Cedar wood should be used to construct the case as this material will be readily acceptable to the bees.

Cedar wood will also withstand the weather without the need for painting but it is well to remember that if the counter is to be used in exposed outdoor conditions, weather protection becomes an important consideration, whereas in laboratory conditions it is not so.

The best compromise for an outdoor installation is a shelter which will keep off the rain.

First of all make all the wooden parts of the case as detailed in Fig. 4.

Now solder the two thin flexible covered wires to the bulb holder tags and screw in the bulb. These wires are led out through the top of the base and the bulb assembly is glued in position.

It is not likely that the bulb will need replacement because it is "under run" and there is a 20 ohm resistor (R1) in series with the bulb which reduces the light and heat dissipated in the bulb.

When the glue has set, fill up the light hole with Perspex cement so that it comes flush with the passage floor.

Glue down the two sides of the tunnel so that the width of the narrowest region is 14 in. Paint the tunnel top, bottom and sides a matt black.

The light dependent resistor should be a push

fit into a hole in the tunnel roof.

Glue and screw the front and back to the base and glue the exit ramp in position. Drop the tunnel roof into position indicated. The other parts of the case are made from Perspex and their dimensions are given in Fig. 4.

With these made we can proceed with the assembly.

ASSEMBLY

Begin by screwing the Perspex side and top windows in position as indicated. Glue the Perspex platform to the front and place the Perspex exit guide in position.

Now solder the two wires from the bulb holder to the component board as detailed in Fig. 2, push the l.d.r. in position and then attach the board to the back of the case by means of a 4 B.A. nut and bolt. This bolt should be countersunk into the back so the back is flush with the front of the hive. If there is a gap here, the bees will try to go in or out through the smallest crevice.

Attach the wander sockets to the Paxolin side and solder to the appropriate flying leads from the component board. Next screw the Paxolin side to the case.

When the flying leads to the counter have been connected, fit the counter into its locating holes, (one end in the Paxolin and the other in the bracket on top of the Perspex exit guide) and secure with nuts. The counter digits should be visible through the slot in the Paxolin side.

Screw the top on and the unit is complete.

CAPACITY AND POSITION OF CASE

The single entry counter (as this is) is only suitable for a three or four frame hive, since with a full scale hive the returning bees would sometimes overload the tunnel capacity.

The maximum a single entry counter can

handle is about 60 per minute.

For a full scale hive a three entry counter is necessary. This means the entry tunnel is divided into three passages, each with its own light beam arrangement, amplifier and counter.

Whereas the single entry model is only 6¹4in, wide, which is about right for most observation hives, it is better to make the three entry model 16¹2in, wide so that it takes up the whole width of a Standard National hive.

When the counter is put in front of the hive the hive should be moved back by a distance equal to the depth of the Bee Counter, in this case 314 in. so that the point of entry is exactly as it was without the counter.

When this is done the bees will soon get used to the new conditions and will be using the exit and entry passages without any confusion.





LOGATOB

to remove unwanted a.c. from the voltmeter input, and diodes D1 and D2 protect the meter movement against overload.

At a certain setting of C4, the d.c. voltage at TR1 emitter will equal the voltage at the junction of R5 and R6 so that no current flows through ME1; this can be taken as the normal operating point for the circuit. If metal is brought close to L1, the emitter voltage of TR1 will rise by several millivolts in relation to the voltage at the junction of R5 and R6, and the meter will read.

Full scale sensitivity of the null voltmeter is around 150 millivolts. Metal Locator response is shown in Fig. 2, where meter reading is plotted against depth for three weights of metal.

CONSTRUCTION

Commence construction by cutting a piece of 0·1 inch matrix plain perforated circuit board to a size of 3·1 by 1·4 inches, and drill holes to take C4, VR1, and S1 (see Fig. 3).

Cut two brackets from a length of 12 inch aluminium angle and drill to accept the meter terminal screws and 6B.A. circuit board mounting screws.

Bolt the brackets to the circuit board, complete with solder tags, and insert all terminal pins in the positions shown in Fig. 3.

With C4, VR1, and S1 in place on the circuit board, proceed to mount and solder the remaining components in the following order; resistors, capacitors, wire links and leads, diodes and the transistor, using a heat shunt to protect the diodes and transistors while soldering them.

Obtain a plastic beaker with lid (of minimum dimensions 5 inches high by 2^{1}_{2} inches diameter) and cut away the centre of the lid to accept the meter MEI. Next, drill holes in the beaker for L1 leads, woodscrews, and to allow access to the circuit board controls, see Fig. 4.

When following the step-by-step instructions in Fig. 5, for making up the search coil L1, ensure that the pile windings can slide easily off the 5 inch diameter former. Short strips of insulating tape, placed sticky side out around the former, will hold the turns together and facilitate removal of the coil. Do not use Sellotape for this purpose as it is likely to damage the wire.

The metal locator frame (Fig. 4 and 6) consists of a chipboard or plywood handle, a ⁵8 inch diameter dowel pole, and two s.r.b.p. or Perspex sheets for the search head. Screw and glue the handle to the pole and then glue the other end of the pole to the search head top board, this assembly can then be painted.

To complete the construction, screw the

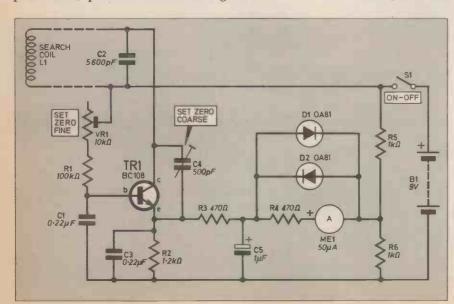


Fig. 1. Circuit diagram of the Metal Locator. The search coil L1 is mounted in the locator head and the dotted lines are the connecting wires to the circuitry.

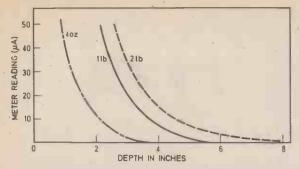


Fig. 2. Response curves of the Metal Locator.

plastic beaker to the pole opposite the handle, securely clamp the search coil between the boards, run twin leads from L1 to the beaker, and position the battery.

In the prototype, the battery was held in place behind the meter with a rubber band, as shown in the photograph, but it could equally well be fixed inside the beaker with a small clip or elastic band.

SETTING UP

Adjust VR1 to mid track, C4 to minimum capacitance (unscrewed), and switch on. The meter pointer should go beyond full scale. With the search coil well away from metal objects, screw in C4 until the meter reads somewhere between zero and full scale. Trim for a zero reading with VR1.

OPERATING LICENCE

The Metal Locator described in this article is designed to operate in the frequency band specified by the Ministry of Post and Telecommunications (16 to 150kHz). The circuit design of the locator should not be altered in any way that may affect the operating frequency.

A licence must be obtained before using the locator; this costs 75p for 5 years. An application form for a licence is obtainable from the Ministry of Post and Telecommunications, Waterloo Bridge House, Waterloo Road, London, S.E.1.

If the meter fails to read, or no response is obtained from adjustment of C4, check for wiring errors.

A certain amount of drift will be evident immediately after the locator has been switched on, therefore allow the circuit to settle down and then readjust C4 and VR1. Locator response can then be checked with metal weights and compared with Fig. 2.

Increased sensitivity can be achieved by reducing the value of C3 to 0·15µF, but this will enhance circuit drift to the point where frequent adjustment of VR1 is necessary. Conversely, drift and sensitivity will be reduced if C3 is increased in value.

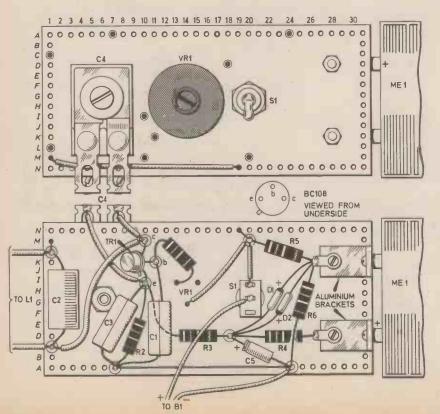


Fig. 3. Top and underside views of the circuit board and meter assembly. The circled connections represent the terminal pins used in the construction of this item. These pins are clearly indicated in the top diagram.



Components....

Resistors

R1 $100k\Omega$

R2 1·2kΩ R3 470Ω

 $R4 470\Omega$

 $R5 1k\Omega$

 $R6 1k\Omega$

All ± 10% ½ watt carbon.

Capacitors

C1 0.22 µF polyester 250V

C2 5,600pF polystyrene

C3 0.22 pF polyester 250V

C4 500pF mica compression trimmer

C5 1µF elect. 12V

Semiconductors

TR1 BC108 silicon npn

D1 OA81

D2 OA81

Meter

ME1 50μA f.s.d. moving coil. SEW type MR 38P

Switch

S1 S.P.S.T. sub-miniature toggle

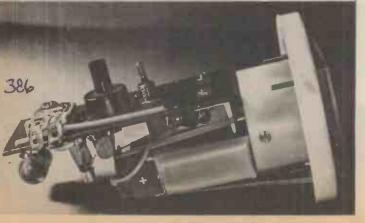
Miscellaneous

VR1 10k Ω miniature carbon T.V. type preset B1 PP3 battery. Circuit board 3·1 inch by 1·4 inch plain, perforated 0·1 inch matrix Veroboard and Veropins. 26 s.w.g. cotton covered or enamelled copper wire, plastic beaker (see text), connecting wire, wood and screws for assembly, $\frac{1}{2}$ in aluminium angle for brackets.

USE

The locator is now ready for use and can be used for beachcombing or searching the back garden or waste ground. The locator may be subjected to damp and the pole, in particular, should be painted for protection if nothing else.

Photograph showing the construction of the circuit board and meter mounted on the beaker lid.



Continued from page 361

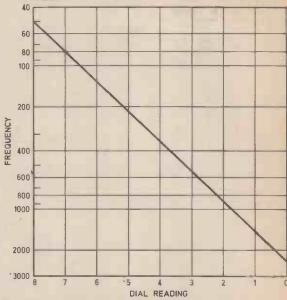


Fig. 8. Approximate output frequency for various control settings.

FINAL ASSEMBLY

Final assembly amounts to attaching the front panel to the box frame with self tapping screws, fitting the battery inside and fitting rear panel.

The generator can be connected to the input of any amplifier but the signal output level should be adjusted in accordance with that required by the amplifier input. To comply with the calibration chart given in Fig. 8 turn VR1 fully anti-clockwise and fix the frequency control knob to read zero. The output control knob is fixed in the same way i.e., to read zero with VR2 fully anti-clockwise.

The Audio Tone Generator is now ready for use and can be tried out in conjunction with a tape recorder.



Everyday Electronics, May 1972

LARGEST SELECTION OF SEMICONDUCTORS COMPONENTS

RETURN OF POST SERVICE

_				TR	ANS	ISTO	RS				
2G301	20p	2N3404	32}p	40310	45p	BC212L	13p	BSX28	32 p	NKT28	1 27 to
2G302	20p	2N3405	45p	40311	35p	BCY30	27 p	BSX60	82 p	NKT40	1 871p
2G303 2G306	20p	2N3414 2N3415	22 dp 22 dp	40312 40314	471p	BCY31 BCY32	30p 50p	BSX 61 BSX 76	62 d p 22 d p	NKT40	2 90p 3 75p
2G308	30p	2N3416	2110	40320	278P	BCY32 BCY33	25 p	BSX77	27 p	NKT40	
2G309 2G371	30p	2N3417 2N3570	37±p	40323 40324	32 p	BCY34	30p 40p	BSX78	27tp	NKT40	5 75p
2G371	15p 20p	2N3570	£1-25 97 p	40324	474p 374p	BCY38 BCY39	60p	BSY10 BSY11	271p	NKT40	6 624p 1 624p
2G381	22 in	2N3605	27 p	40329	30p	BCY40	50p	BSY24	15p	NKT455	2 62 p
2N404 2N696	221p 20p	2N3606 2N3607	271p	40344	27ip	BCY42 BCY43	15p 15p	BSY25 BSY26	15p	NKT45	3 4710
2N697	17p	2N3702	221p	40348	591n	BCY54	321 p	BSY27	174p	NKT612	3F 32+p
2N698	25p	2N3703	10p	40360	421p	BCY58 BCY59	221 p	BSY28	17#P	NKT67	4F 30p
2N706 2N705A	12ip 12ip	2N3704 2N3705	11p 10p	40361 40362	471p	BCY59	221p 971p	BSY29 BSY32	174p 25p	NKT677	7F 30p 3 25p
2N708	15p	2N3706	09p	40370	32 p	BCY70	20p	BSY36	25p	NKT78	1 30p
2N709 2N718	62 jp 25 p	2N3707 2N3708	11 p 07 p	40406 40407	57 p	BCY71 BCY72	25p	BSY37 BSY38	25p	NKT10- NKT10-	419 30p
2N726	30p	2N3709	09p	40408	52 p	BCZ10	17 p 27 p	B8 Y39	22 lp 22 lp	NETIU	871p
2N727	30n	2N3710	09p	40410	624p	BCZII	421p £1.121	BSY40	32 i p	NKT10	519
2N914 2N916	171p	2N3711 2N3715	12p £1.25	40467A 40468A	57½p 35p	BD116 BD121	£1.12}	BSY51 BSY52	32 p	NKT20	32}p
2N918	30p	2N3716	£1.30	40600	574p	BD123	821p	BSY53	37 p 40p		47 tp
2N929	221p	2N3791	£2.06	AC 107	574p	BD124	60p	B8Y54		NKT20	339
2N930 2N1090	271p	2N3819 2N3823	35p 97∳p	AC126 AC127	20p 25p	BD131 BD132	75p 85p	BSY56 BSY78	90p	NKT801	37≩p
2N1091	22 ip	2N3854	271p	AC128	20p		£1-37	BSY79	45p	11 16 1 001	771p
2N1131	25p	2N3854	4 274p	AC154	224p	BDY11	21-62	B8 Y82	521p	NKT801	112
2N1132 2N1302	25p 174p	2N3855 2N3855	271p	AC176 AC187	25p 624p	BDY17 BDY18	£1.50 £1.75	BSY95 A	574p	NKT801	97∦p
2N1303	174p	2N3856	30p	AC188	87èn	BDY19	£1.97}	B8W41	42 P		£1.12
2N1304	22 p	2N3856/		ACY17 ACY18	27 p	BDY20 BDY38	£1.12#	BSW70	27 ip	NKT80:	211 92≟p
2N1305 2N1306	22½ p 25p	2N3858 2N3858	25p	ACY19	25p 25p	BDY60	97 p 21.25	C111 C424	75p 27ip	NKT802	212
2N1307	25p	2N3859	27 p	ACY20	25p	BDY61	£1.25	C425	55p		92}p
2N1308 2N1309	30p 30p	2N3859/ 2N3860	324p	ACY21	25p 20p	BD Y62 BF115	£1.00 25p	C426 C428	40p 37↓p	NKT802	921p
2N1507	17 p	2N3866	£1.50	ACY28	20p	BF117	471p	C744	30p	NKT802	214
2N1613	25 p	2N3877	40p	ACY40	20p	BF163 BF167	374p	D16P1	37 p	NT YE STOOM	92 ∤ p
2N1631 2N1632	35p 30p	2N3877 A 2N3900	37ip	ACY41 ACY44	25p 40p	BF173	18p 19p	D16P2 D16P3	40p 37∮p	NKT802	921p
2N1638	274p	2N3900/	40p	AD140	521n	BF177	30p	D16P4	40p	NKT802	216
2N1639 2N1671B	27 p	2N3901	97ip	AD149	57èp	BF178	30p	GET102 GET113	30p 20p	OC20	921p 75p
2N1711	25n	2N3903 2N3904	35p 35p	AD150 AD161	62 p	BF179 BF180	30p 35p	GET114	20p	OC22	50p
2N1889	32}p	2N3905	371p	AD162	37 p	BF181	321p	GET118	20p	OC23	60p
2N1893 2N2147	37 p	2N3906 2N4058	374p	AF106 AF114	42 p 25 p	BF184 BF185	25p 42ip	GET119 GET120	20p	OC24 OC25	60p 50p
2N2148	57 p	2N4059	10p	AF115	25 p	BF194	17 p	GET873	524p 124p	OC26	271p
2N2160	57†P	2N4060	1210	AF116	25p	BF195	15p	GET880	30p	OC28	62 ln
2N2193 2N2193A	40p	2N4061 2N4062	121p	AF117 AF118	25p 62 p	BF196 BF197	42½p 42½p	GET887 GET889	20p 22}p	OC29 OC35	62 i p 50p
2N2194A	30p	2N4244	47}p	AF119	90m	BF198	491 n	GET890	224 P	OC36	8010
2N2217 2N2218	27 p	2N4285	17}p	AF124 AF125	22 i p 20 p	BF200 BF224	52 p	GET896 GET897	221p 221p	OC41 OC42	22ip 25p
2N2218 2N2219	23p 23p	2N4286 2N4287	171p 171p	AF126	20p	BF225	14p 19p	GET898	221P	OC44	20n
2N2220	25p	2N 4288	174p	AF127	17∤p 37∤p	BF237	23p	MJ400	£1.07	OC45	12}p
2N2221 2N2222	25p 30p	2N4289 2N4290	174p	AF139 AF178	37 p	BF238 BF244	23p 23p		£1.12}	OC46 OC70	15p 15p
2N2270	47 t D	2N4291	17 p	AF179	72 t p	BFW61	471p	MJ430	£1.02}	OC71	124p
2N2297	30p	2N4292	12 jp	AF180	524p	BFX12	22 i p	MJ440	95p	OC72	TX#D
2N2368 2N2369	17 p 17 p	2N4303 2N5027	471p	AF181 AF239	421p 421p	BFX13 BFX29	224p 30p	MJ480 MJ481	97 p 21.25	OC74 OC75	321p 221p
2N2369A	1710	2N5028	57 t p	AF279		BFX30	30p	MJ490	£1.00	OC75 OC76	22 + p
2N2410 2N2483	42 p	2N5029	471p	AF280 AF211	82ip	BFX42 BFX44	371p		21.37	OC77 OC81	30p 20p
2N2484	27 p 32 p	2N5030 2N5172	42 lp	A8Y26	25n	BFX68	87 p	MJE340	£2-17} 62}p	OC81D	22 tp
2N2539	22 p	2N5174	52 p	ASY27	371p	BFX84	25p	MJE520	60 p	OC83	25p
2N2540 2N2613	22∳p 35p	2N5175 2N5176	52 lp 45p	ASY28 ASY29	27 p	BFX85 BFX86	32 p 25 p	MJE521 MPF102	73 p 42 ½ p	OC84 OC139	25p 321p
2N2614	30p	2N5232A	30p	ASY36	25p	BFX87	27 p	MPF103	37 p	OC140	32 i p
2N2646	521p	2N5245	45n	ASY50	25p	BFX88	25v	MPF104	371p	OC170	30p
2N2696 2N2711	32 p 25p	2N5246 2N5249	42 p 67 p £3.25 £2.75	ASY51 ASY54	32 t p 25 p	BFX89 BFX93A	621p 70p	MPF105 MP83638	371P	OC171 OC200	30p 40p
2N2712	25p	2N5265	£3.25	A8¥86	32 ip	BFY10	32 + p	NKT001 NKT124	3 47 P	OC201	60p
2N2713 2N2714	27ip	2N5266 2N5267	£2.75	AU103 ASZ21	£1.25	BFY11 BFY17	424p 221p	NKT124 NKT125	42 p 27 p	OC202 OC203	75p 42+p
2N2865	62}p	2N 5305	37 p	BC107	10n	BFY18	321p	NKT126	27 t p	OC204	42 ł p
2N2904 2N2904A	30p 32∤p	2N5306 2N5307	40p	BC108 BC109	10p 10p	BFY19 BFY20	32 tp	NKT128 NKT135	271p	OC205 OC207	90p 75p
2N2905	37 p	2N5308	37 p	BC113	15p	BFY21	421p	NKT137	27 p 32 p	OCP71	42 p
2N2905A	40p	2N 5309	62 ł p	BC115	15 p	BFY24	45p	NKT210	30p	ORP12 ORP61	50 p
2N2906 2N2906A	25p 27+p	2N5310 2N5354	421p 271p	BC116A BC118	15p 10p	BFY25 BFY26	25p 20p	NKT211 NKT212	30p	P346A	50p 22ip
2N2907	27 p 30p	2N5355	27 p	BC121	20p	BFY29	50p	NKT213	30p	T1834	624B
2N2923 2N2924	15p 15p	2N5356 2N5365	32 p	BC122 BCI25	20p 20p	BFY30 BFY41	50p 50p	NKT214 NKT215	224p	TIS43 TIS44	27p 10p
2N2925	15p	2N5366	471p 321p	BCI26	20p	BFY43	62 p	NKT216	374P	T1845	10p
2N2926		2N5367	57 p	BC140	37 p 10 p	BFY50	23p	NKT217	424p	TI846 TI847	11p
Green Yeilow	14p 124p	2N5457 28005	75p	BC147 BC148	10p	BFY51 BFY52	20p 23p	NKT219 NKT223	30p 27}p	T1848	11p 12}p
Orange	12}p	28020	£2.00	BC149	12p	BFY53 BFY56A	171p	NKT224 NKT225	25p	T1849	121n
2N3011 2N3014	30n	28102 28103	50p 25p	BC152 BC157	17ip 20p	BFY56A BFY75	571p 30p	NKT225 NKT229	22 i p 30p	T1850 T1851	17ip 12ip
2N3014 2N3053	321p	28103	25p	BC157	11p	BFY76	424p	NKT237	35p	T1852	124p
2N3054	46p	28501	324 p	BC159	12p	BFY77	57 p	NKT238	25p	T1853	22 † p
2N3055 2N3133	62p 30p	28502 28503	35p 27ip	BC160 BC167	62}p	BFY90 BFW58	67èp 27èp	NKT240 NKT241	271p 271p	T1860 T1861	22 p 25 p
2N3134	30p	3N83	40p	BC168B	10p	BFW59	25p	NKT241 NKT242	20p	TI862	27 p
2N3135	25p	3N128	70p	BC168C	11p	BFW60	25p	NKT243	624p	TIP29A TIP30A	50p
2N3136 2N3390	25p 25p	3N140 3N141	771p 721p	BC169B BC169C	11 p 12 p	BPX25 BPX29	£1.85 £1.80	NKT244 NKT245	17 p 20p	TIP30A	60p 62}p
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Last month we posed some problems under the heading Teach in Half-Term Test. We will now answer those problems and try to show how we arrived at the answers. If you have got some of them wrong do not worry, just try and follow our explanation and see where you went wrong.
(1) They flow from negative to positive in reality. Al-

though we assume that conventional current flows from positive to negative the actual electrons flow from negative to positive.

(2) (b) μ A (microamps), (e) A (amps) (3) 22 volts. V = IR hence V = $0.01 \times 2.2 \times 1,000 = 22$ V (4) It does not matter. All the resistor does is to limit

the current; this can be done at any point around the circuit.

$$R_{T} = R1 + \frac{R2 \times R3}{R2 + R3} = 10 + 33.3 = 43.39$$

the circuit.
(5) $2\cdot 8mA$. Total resistance is $2\cdot 2k\Omega + 1k\Omega = 3\cdot 2k\Omega$.

Current flow $I = \frac{V}{R} = \frac{9}{3\cdot 2\times 1,000} = 2\cdot 8\div 1,000A = 2\cdot 8mA$ (6) R1 and R3 $\frac{1}{2}$ W, R2 1W. Total circuit resistance $R_T = R1 + \frac{R2\times R3}{R2+R3} = 10 + 33\cdot 3 = 43\cdot 3\Omega$.

Total current $I = \frac{V}{R} = \frac{9}{43\cdot 3} = 0\cdot 21$ A

Dissipation of R1 = $I^2R = 0\cdot 21\times 0\cdot 21\times 10 = 0\cdot 44$ W. The nearest commercial rating $I = \frac{1}{2}$ W. Next calculate the voltage drop across R2 and R3 together $V = IR = 0\cdot 21\times 10^{-1}$ voltage drop across R2 and R3 together V = IR = 0.21 ×

33.3 = 7V.We know that $W = I^2R$, but $I = \frac{V}{R}$ therefore

$$W = \frac{V}{R} \times \frac{V}{R} \times R$$
 and, cancelling $W = \frac{V^2}{R}$

W =
$$\frac{V}{R} \times \frac{V}{R} \times R$$
 and, cancelling W = $\frac{V^2}{R}$

Dissipation in R2 = $\frac{V^2}{R} = \frac{7 \times 7}{50} = \frac{49}{50} = 0.98W$

Dissipation in R3 = $\frac{V^2}{R} = \frac{49}{100} = 0.49W$

(7) 0.4W or 400mW Maximum dissipation occur

(7) 0.4W or 400mW. Maximum dissipation occurs when the value of VR1 equals that of R1 i.e. 50Ω . When both resistors are of equal value the voltage drop across each is half the voltage drop across both, therefore, maximum dissipation in VR1

 $= \frac{V^2}{R} = \frac{4.5 \times 4.5}{50} = \frac{20.25}{50} = 0.405W$

- (8) (a) $4.7k\Omega \pm 10\%$
 - (b) $22k\Omega \pm 5\%$ (c) $100 \text{k}\Omega \pm 10\%$

(9) (b) 20μF 40V. In most applications using electrolytic capacitors the capacitance must be greater than a certain value; the tolerance of a normal 16µF would encompass 20µF. The important thing is that the working voltage is the same or greater.

(10) Reject it politely. He has given you a 120,000pF or 0·12μF capacitor. Check to see If he has the precise value and, if he does not, you may as well take this one, since it should be near enough to use as a substitute.

(11) C1 will charge up the fastest as it has the lowest value and is being charged through the lowest value resistor.

(12) C2 will take the longest time to charge, as it has the highest value and is being charged through the highest value resistor.

(13) Forward biassed. The conventional current flows from positive to negative and can thus flow through

the diode in the direction of the arrow.

(14) 100V and 100mA. Peak reverse breakdown voltage will be the battery voltage. Since in the reversed blassed condition there is negligible current flowing R1 will not drop any voltage and the full supply voltage will appear across D1. In the forward biassed condition the dlode can be assumed to be a short circuit thus only R1 can limit the current flowing hence

$$I = \frac{V}{R} = \frac{100}{1 \times 1,000} = 0.1 \text{A or } 100 \text{mA}$$

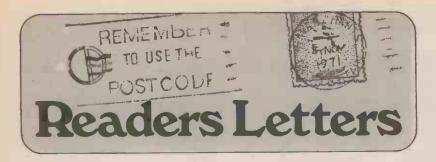
(15) (d) 100V, 150mA. Both ratings given are minimum ratings, 0.1A = 100mA.

(16) (b) 0.6V. As the diode is forward biassed the voltage would be 0.6V. There is always a voltage drop of approximately 600mV across silicon diodes due to the "knee" in the characteristic.

Well, how did you fare? If you got them all right that is excellent, if you did not the important thing is that you understand where you had difficulties. We suggest that you re-read the relevant sections of the Teach-In series.

We hope that you found the questions a challenge and at the same time they have opened your eyes to some calculation methods-particularly the calculation of dissipation. If you used $W = I^2R$ Instead of deriving $W = \frac{1}{R}$

this does not matter but it may pay to look for an easier way next time.



Bias Value

Having been a subscriber to P.E. and P.W. "off and on" for about 10 years I came across the January issue of EVERYDAY ELECTRONICS, which had my instant approval and now joins the rank of my other magazine's culminating in an endless and very informative pile on top the piano.

I find it is a magazine not only of theoretical enthusiasm but of great practical interest to the "everyday handyman" and certain to be a book for beginners, especially the very helpful facts "projected" by Mike Hughes, M.A.

I would hope in the future that perhaps Mr. Hughes could give reference to finding values of bias resistors, etc., needed for the satisfactory operation of different transistor parameters, and also relevant circuit operation of thyristors, unijunction and field effect transistors and other very useful flexible types of semiconductors.

Noticing other readers' troubles referring to the *Electro Laugh*, I also constructed this article and it worked first time owing to the way I adopt when working on, or constructing any project, I always check the finished article with the actual circuit diagram thus finding our little friend Q7 and P7.

Unfortunately the only earphone I had was a high impedance crystal type, but by connecting a resistor in the region of 250 ohms in parallel with it, it brought the overall impedance down to a satisfactory level with a slight reduction in volume.

J. Mason S. Wales

We doubt if Teach-In will be able to meet all your needs as it will finish after 12 months. However we will be publishing further series that should help.

Another Bug

Naturally, I was quite flattered to discover that you had found my letter sufficiently interesting for inclusion in Readers Letters (March issue), however, I must admit that my pleasure was mixed with large helpings of disappointment and frustration due to your editing of the letter.

I am not complaining at all about the amount of space allocated to my comments—I realise you have the right to include only that which in your wisdom you decide is worthy of publication.

My complaint is that you have entirely neglected to make even a brief reference to what was after all the main point of my letter-the difficulty of obtaining items advertised in your magazine. By omitting any reference to this frustrating situation, my letter as printed is sailing under false colours-the few minor constructional queries were in fact, sorted out by trial and error once I got going. The real reason for being unable to get cracking was not so much mounting components, as actually getting hold of them!

The fact that you completely ignored my comments regarding suppliers leads me to two conclusions:

(One) That you accepted my comments to be an exaggeration of a somewhat hysterical nature, and were not a true picture of the real situation, or

(Two) That you accepted my statements as correct, but did not wish to offend your advertisers whose business you must obviously wish to retain.

With regard to the former, I feel I, must now justify my remarks by quoting a few of the more deplorable examples of SERVICE, and leave you to form your own conclusions. These examples are on a separate sheet herewith enclosed.

Regarding (Two), whilst I

realise that you are not to be held responsible for goods or advertised in your services columns, you do, however, have a moral responsibility to your readers. After all, it is you that place these offers before us, the readers, and if for example, I had not seen a certain item offered in your magazine, then I would have been saved the trouble and frustration that followed when the item failed to arrive, and all attempts to obtain satisfaction are largely ignored.

However, I have now found a couple of very good suppliers whose friendly, courteous, and extremely efficient service have allowed me to obtain some of the pleasure that I had hoped would be derived from my new hobby (Galleon Trading Co. and

Radio Exchange Co.).

To date I have completed several very efficient radios, some from kits; also the Astron, a general purpose amplifier, and one or two other gadgets, and success rate so far is quite satisfactory, so the situation is not too black after all.

J. G. Richards Sale, Cheshire

The above correspondent supplied us with details of orders placed with four different advertisers, none of which had been expediently dealt with, at the time of writing.

We have investigated all of these cases on behalf of our reader. The delays, regrettable as they are, seem to be unavoidable and can be largely attributed to the phenomenal success of this magazine's declared intention to popularise the hobby of electronics!

As a consequence, our advertisers are sometimes overwhelmed by a flood of orders, and delays do therefore sometimes arise. But we know all our advertisers make determined efforts to clear their back-log of orders as quickly as possible.

We, on our part, will always investigate any serious and reasonable complaints, on behalf of our readers.

Cell Life

I have just read the March issue of EVERYDAY ELECTRONICS and thoroughly appreciated the Ruminations by Sensor where he mentioned the tin saw and how much damage could result to a

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500mA (class B working). Takes the place of any
of the following batteries: PPI, PPS, PP4, PP6,
PP7, PP9 and others. Kit comprises: mains
transformer rectifier, smoothing and load resistor.
condensers and instructions. Real snip at only
88p, plus 20p postage.

5 amp. changeover contacts, 9p each, \$1 doz. 15 amp. Model 10p each or \$1.05 doz.



EXTRACTOR FAN Cleans the air at the rate of 10,000 cubic fiv. per hour. Suitable for kitchens, bathrooms, factories, changing rooms, etc., it's so quiet it can hardly be heard. Compact, 5\(\frac{1}{2}\) casing with 5\(\frac{1}{2}\) fan blades. Kit comprises motor, fan blades, sheet steel casing, pull switch. mains connector, and fixing brackets, 22 plus 38p poet and ins.

MAINS MOTOR

post and ins

Precision made—as used in record decks and tape recorders-ideal also for extractor fan, blower, heaters, etc. New and perfect. Snip at 50p. Postage 15p for first one then 5p for each one ordered.



Continuously variable 30°-90° C. Has sensor bulb connected by 33in. of flexible tubing. On operation a 15 amp 250 volt switch is

opened and in addition a plunger moves through approx in. This could be used to open valve on ventilator etc. £1.50 plus 23p p. & ins

5A 3-PIN SWITCHED SOCKETS

An excellent opportunity to make that bench dis board you have needed or to stock up for future jobs. This month we offer 6 British made (Hjersit) bakelite flush mounting shuttered switch sockets for only 50p plus 18p post and insurance (10) became not from ance. (20 boxes post free).



MAINS OPERATED

SOLENOIDS Model 772—smail but powerful 1° pull—approx. size
1½° x 1½° x 1½° x 1½° 60y.
Model 400/1 ½° pull. Size
2½° x 2° x 1½° 75p.
Model TT10 1½° pull. Size 3° x 2½° x 2½° £1:80

20p post and insurance.

A TELESCOPIC AERIAL

for portable, car radio or transmitter. Chrome plated— six sections, extends from 7½ to 47in. Hole in bottom for 6BA screw. 38p. KNUCKLED MODEL FOR F.M. 56p.



This Tuner is a precision instrument made by the famous Cyldon Company for the equally famous Radiomobile Car Radio. It is a medium wave tuner (but set of longwave coils available 25p) with a frequency coverage 1620 Kc/s-526 Kc/s and intended to operate with an LF. value of 470 Kc/s. Extremely compact (size only 2½ × 2 × ½ ins. thick) with reduction gear for fine tuning. Bnip price this month 50p, with circuit of front end suitable for car radio or as a general purpose tuner for use with Amplifier. Post free.

RADIO STETHOSCOPE

Easiest way to fault find—traces signal from aerial to speaker—when signal stops you've found the fault. Use it on Radio, TV,

santi- Ose to on Radio, TV, amplifier, anything—complete kit comprises two special transistors and all parts including probe tube and crystal earpiece. \$2—twin stethoset instead of earpiece 75p extra post and ins. 20p.

STANDARD WAFER SWITCHES -

Standard size 14 wafer—silver-plated 5-amp contact, standard 2 apindle 2 long—with locking washer and nut.

No. of Poles	2 way	3 way	4 way	5 way	6 way	8 way	9 way	10way	12way
1 pole	40p	40p	40p	40p	40p	40p	40p	40p	40p
2 poles	40p	40p	40p	40p	40p	40p	40p	70p	70p
3 poles	40p	40p	40p	40p	70p	70p	70p	95p	95p
4 poles	40p	40p	40p	70p	70p	70p	70p	£1 ·20	#1 .20
5 poles	40p	40p	70p	70p	95p	95p	95p	£1 ·45	£1 ·45
6 poles	40p	70p	70p	70p	95p	95p	95p		
7 poles	70p	70p	70p	95p		£1 -20		£1 -95	
8 poles	70p	70p	70p	95p		\$1.20			
9 poles	70p	70p	95p	95p		21 -45			
10 poles	70p	70p	95p	£1 -20		£1 ·45			
11 poles	70p	95p	95p			£1 70			£2 ·95
12 poles	70p	95p	95p	£1 ·20	#1 -70	£1·70	£1·70	£3 ·20	£8·20

THYRISTOR LIGHT DIMMER

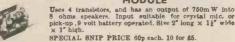
For any lamp up to 200 watt. Mounted on switch plate to fit in place of standard switch. Virtually no radio interferences. Price £1.99 plus 20p post and insurance.



- THIS MONTHS SNIP

1 HOUR MINUTE TIMER. Made by Smiths complete with control knob and calibrated dial. This month's special bargain at 50p. Useful in the Kitchen, Office and Dark-room etc.

MULLARD AUDIO AMP LIFIER MODULE



POCKET CIRCUIT TESTER

Test continuity for any low resistance circuit, house wiring, car electrics. Test polarity of diodes and rectifiers. Also ideal size for conversion to signal injector (circuit supplied), 30p or 2 for 50p. Post paid.



METAL LOCATOR AUDIO TONE GENERATOR BEE COUNTER

To receive details on these kits send s.a.e. for parts list.

MULLARD I.F. MODULE

This is a fully screened intermediate frequency module amplification and detection of f.m. signals at 10-7MHz and am. signals at 470kHz. The first stage is used as an i.f. amplifier for f.m., and a set? oscillating mixer for am. operation, in conjunction with an external oscillator coil. 75p each. 10 for \$6.75.



DISTRIBUTION PANELS

Just what you need for work bench or lab.
4 x 13 amp seckets in metal box to take
standard 13 amp fused plugs and on/off switch with neon warning light. Supplied
complete with 6 feet of flex cable. Wired up ready to work, 22:25 plus 23p P, & I

BATTERY CONDITION TESTER

Made by Mailory but suitable for all batteries made by Kver Ready and others, most of which are zinc carbon types but also hercury manganese—nicad—silver oxide and alkaline batteries may be tested. The tester puts a dummy load on the battery and the meter scale indicates the condition depending upon which section the pointer rests. The section reads "replace" "weak" or "good". The tester is complete in its case, size 3% x 6% x 2" with leads and prods. Price \$1.75 plus 20p postage.

Where postage is not stated then orders over £5 are post free. Below £5 add 20p. Semiconductors add 5p post. Over £1 post free, S.A.E. with enquiries please.

Thermostat with Probe. Made by the famous Ranco Thermostat Co. Covers the range from approx. ©-20°0C. variable by a control spindle, handles currents up to 16 amps. Length of capillary and sensor tube approx. 3° 6°. These are ideal for ovens and as a general purpose thermostat. Price 50p each or 10 for \$4.50. Small Tuning Condenser as fitted to many imported Japanese and Hong Kong radios. 2 gang about 200PF per gang. Size approx. 1° x 1° with a \$\frac{1}{2}\$ diameter spindle with dust cover. 25p each or 10 for \$2.25. Beat Sink. Small type as used with OC81 etc. Price 5p each or 10 for \$2.25. Beat Sink. Small type as used with built-in hearing aids. The amplifier and battery being housed in the arms. Although these are complete hearing aids we are selling them purely for the sub miniature components they contain. We give no guarantee that they are in working order also these may be secondand. Price \$25.60 each. Foot Switch. Twin levers each of which operates a 10 amp QMB changeover switch. Price 90p each. Programmers. 5 Revs per minute. Made by Magnette Devices Ltd. The contacts may be set to trigger anywhere around the shaft, ideal for motivated lighting displays, sequential switching etc. Drive motors are 200-240v 50Hz. Model Bhas 11 change over contacts. Price \$2.50 each. The contacts are opper clad? Tubular construction. See the content of the co

Sacker Model 7DJ MMLI again 2000 water rate: but 230-240v, has no cover over clement ends Price 65p each or 10 for 55-85. Slide Switch. 2 pole change over panel mounting by two 6 BA screws. Slze approx. 1° x 1° rated 250v lamp, 6 p each. 10 for 54p. 100 for 25-10. 500

for £24.

As above but for printed circuit 5p each, 10 for 45p, 100 for £4.25.

Bub Ministure Slide Switch. DPDT 19mm (#* approx.) between fixing centres. 12p each or 10 for £1.08.

KITS FOR PREVIOUS PROJECTS-

Unless otherwise stated, kits contain electronic parts only. The case and special items can be obtained locally. Also batteries are not included. Kits may be returned for refund if construction has not been started. We reserve the right to substitute components should deliveries be protracted so as to avoid undue delay.

HOME SENTINEL INTRUDER ALARM Electronic Components with suitable ca 23.75

SNAP INDIGATOR
WINDSCREEN WIPER CONTROL
Components including metal for chassis \$2
RECORD PLAYER. All components, but not
case, loudspeaker, record deck or pick-np
\$5.50 DEMO DECK 26 75 POST PAID PHOTOGRAPHIC COLOUR
TEMPERATURE METER . 22 65 TEMPERATURE METER 2.9 65
ASTRON RADIO ... 23
REMOTE TEMPERATURE ... 25
COMPARATOR ... 42-25
ELECTRO LAUGH ... 22
RAMBISTOR MICROPHONE ... 1-70
AUTO ALERT
All electronic parts and metal bracket 22-50
RAIN WARNING ALARM
All electronic parts and chassis ... 51-80
WA-WA PEDAL ... 22-90
DAREROOM TIMER ... 24-50
SIMPLE CALCULATOR ... 30
SIMPLE CALCULATOR ... 25
BOLL MOSTURE METER ... 23 00
SIMPLE CALCULATOR ... 25
BOLL MOSTURE WETER ... 25
BOLL POWER SUPPLY ... 25
BOLL POW

Mains Transformer, Primary 240v. tapped 220v. Secondary 20v. ; amp. Price 60p each or 10 for 25-40.

for 25-40. Dial Thermometer—reading from 200-525F used on Tricity and other cookers. This has a flange and can be mounted through a 1½" hole or alternatively it can just be rested on the object whose temperature, it is required to measure. Size 2 × ½" overall diameter. Depth ½" below and ¾" above mounting panel. Price 80p each or 10 for 27-20p.



24-HOUR TIME SWITCH

SWITCH
Made by Smiths, these are
AC mains operated, NOT
CLOCKWORK. Ideal for
mounting on rack or shelf
or can be built into box
with 13A socket, 2 comperiods per 24 hours, 5 amp changeover contacte
will awithe direction on ord during these periods.
22 50 post and ins. 23p. Additional time contacts
50p. pair.

BULL (ELECTRICAL

(Dept. E.E.) 7 Park Street, Croydon CRO IYD Callers to: 102/3 Tamworth Road, CROYDON

HENRY'S LOW COST FIRST GRADE BRANDED GERMANIUM and SILICON TRANSISTORS, DIODES, RECTIFIERS, NEW BRANDED TRIACS, FETS, ZENERS, BRIDGES and INTEGRATED CIRCUITS BY ATES - EMIHUS - FAIRCHILD - FERRANTI - I.T.T. - MULLARD - NEWMARKET - PHILIPS - R.C.A. - TEXAS

A SELECTION FROM OUR LIST

AAY42	10- 1			
AAY42 AAZ13	10p	BD115 75p]	OC16 50p j	2N1303 18p
AAZ13	15p	BD123 85p	OC20 85p	2N1304 22p
	10թ	BD124 80p	OC22 50p	2N1305 22p
AAZI5	10p	BD131 75p	OC23 60p	2N1306 25p
AAZ1/	10p 35p	BD132 80p BDY11	OC24 60p OC25 40p	2N1307 25p
AC126	30p	£1.50	OC26 25p	2N1308 25p
AC127	25p	BDY17	OC28 60p OC29 60p	BM1200 05-
AC127 AC127Z	50p	£1.50	OC29 60p	2N1613 20p 2N2147 75p
AC128	20p	BDY38 65p	OC35 50p	2N2147 75p
AC176	20p	BDY60	OC36 60p	2N2160 60p
AC187	25p	£1.00	OC42 40p	2N2217 25p
AC188 :	25p	DE115 050	OC43 50p OC44 15p	2N2218 20p 2N2218A
ACY17 ACY18 ACY19 ACY20 ACY21	30p 25p	BF115 25p BF154 20p BF158 15p	OC44 15p OC45 12p	2N2218A 25p
ACV19	25p	BF158 15p	OC70 12p	2N2219 20n
ACY20	20p	BF159 35p	OC71 19n	2N2219A
ACY21	20p	BF180 35p	OC72 20p	25p
ACTY 22	10n 1	BF194 15p	OC75 25p	2N2220 25p
ACY39	55p	BF195 15p	OC76 25p	2N2221 20p
ACY40	20p	BF196 15p	OC77 40p	2N2221A
ACY41 ACY44	15p	BF197 15p BFX13 25p	OC81 20p OC81D 20p	25p 2N2222 20p
AD140	25p 50p	BFX13 25p BFX29 25p BFX30 25p	OC81D 20p OC81Z 40p	2N2222 20p
AD149	50p	BFX 30 25p		25p
ADIGI	35p	BFX84 25p	OC84 25p	2N2369 15p
AD162	35p	BFX85 30p	OC139 25p	2N2369A
AF114	25p	BFX84 25p BFX85 30p BFX86 25p BFX87 25p	OC140 40p	15n
AF115	25p	BFX87 25p	OC141 60p	19N9646 40n
AF116	25p	BFX88 20p	OC170 25p	2N2904 20p 2N2904A
AF117	20p	BFY18 25p	OC171 30p	2N2904A
AF118 AF124	60p 25p	BFY50 20p BFY51 20p	OC200 40p OC201 75p	25p 2N2905 25p
A 121.05	20p	BFY51 20p BFY52 20p BFY53 15p BFY90 65p	OC201 75p OC202 80p	2N2905 A
AF126	20p	BFY53 15p	OC202 80p	25p
AF126 AF127 AF139	20p	BFY90 65p	OCP71 97n	2N 2906 20p
AF139	30p	B8 X 20 15p	ORP12 50p	2N2906A
A FIRO	50p	B8X21 20p	ORP60 40p	25n
AF181	45p	BSV27 15n	ST140 15n	2N2907 23p
AF185 AF186	50p	B8T95 A 12p B8Y95 12p	ST141 20p	2N2907A
AF180 AF239	40p	TOTTION 89.9E	TIP29A 50p TIP30A 60p	25p 2N2925 15p
	25p	BY100 15p	TIP31A 60p	2N2996 10n
ASY27	30p	BY126 12p	TIP31A 60p TIP32A 70p	2N3011 20p 2N3053 20p 2N3054 50p 2N3055 75p
ASY28	25p	BY124 15p	TIP33A	2N3053 20p
A8Y29	30p	BY182 90p	£1.00	2N3054 50p
ASY26 ASY27 ASY28 ASY29 ASZ21	55p	BVZ10 85-1	TIP34A	2N3055 75p
BA100 BA102	15p	BYZII 300	21.50	2N3525
BA102 BA110	30p	BYZ12 30p BYZ13 25p	TI843 40p TI850 12p	£1.00 2N3614 55p
BA115	25p 7p	BZY78£1.00	TI850 12p TI851 10p	2N3614 55p 2N3702 10p
BAX13	5n	GET102 35p	TI852 10p	2N3702 10p 2N3703 10p
BAX13 BAX16	5p 7p	GET103 25p	TI860 18p	2N3704 12n
BAY31 BAY38	7 p	GET111 45p	TIS61 20p	2N3705 10p
BAY38	15p	GET113 05%	TI862 25p	2N3707 12p
BC107	10p	GET114 20p	V405A 25p	2N3708 10p
BC108 BC109	10p	GET114 20p GET115 50p GET116 55p GET880 45p GET882 30p	VR525 35p WO1 30p	2N3714
BC109	10p	GET116 55p GET880 45p		22·00 2N3715
BC109C BC113	12p	GET882 30p	WO6 45p	
LICARO			ZTX 107 15n	
BC114	20p	GEX66	ZTX107 15p ZTX108 12p	£2-20
BC114 BC115	20p 20p	£1.50	ZTX107 15p ZTX108 12p ZTX109 15p	£2·20 2N 3716 £2·85
BC114 BC115 BC116	20p 20p 20p	£1.50	ZTX107 15p ZTX108 12p ZTX109 15p ZTX300 12p	£2·20 2N3716 £2·85 2N3771
BC114 BC115 BC116 BC116A	20p 20p 20p 25p	#1.50 GM378A 55p MAT101 25p	ZTX107 15p ZTX108 12p ZTX109 15p ZTX300 12p ZTX301 15p	£2·20 2N3716 £2·85 2N3771 £2·00
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BC114 BC115 BC116 BC116A BC117 BC118	20p 20p 20p 25p 20p 20p 20p 30p 20p	GEX68 GM378A 55p MAT101 25p MAT121 25p MC724P 80p MJ420 80p MJ421 80p MJ2801	ZTX107 15p ZTX108 12p ZTX109 15p ZTX300 12p ZTX301 15p ZTX302 20p ZTX302 20p ZTX304 25p ZTX500 15p ZTX501 15p	£2.20 2N3716 £2.85 2N3771 £2.00 2N3772 £2.05 2N3773 £2.35 2N3791
BC114 BC115 BC116 BC116A BC117 BC118	20p 20p 20p 25p 20p 20p 20p 30p 20p 15p	GEX68 GM378A 55p MATI01 25p MAT101 25p MC724P 80p MJ420 80p MJ421 80p MJ2801 g1:10	ZTX501 15p ZTX502 20p	£2.20 2N3716 £2.85 2N3771 £2.00 2N3772 £2.05 2N3773 £2.35 2N3791 £2.75
BC114 BC115 BC116 BC116A BC117 BC118	20p 20p 20p 25p 20p 20p 30p 20p 15p 20p 20p	#1.50 GM378A 55p MAT101 25p MAT121 25p MC724P 60p MJ420 80p MJ421 80p MJ421 80p MJ2801 #1.10	ZTX501 15p ZTX502 20p	£2.20 2N3716 £2.85 2N3771 £2.00 2N3772 £2.05 2N3773 £2.35 2N3791 £2.75 2N3819 35p
BC114 BC115 BC116 BC116A BC117 BC118 BC119 BC134 BC135 BC136 BC137 BC138	20p 20p 20p 25p 20p 20p 30p 20p 20p 20p 20p 20p	#1.50 GM378A 55p MAT101 25p MAT121 25p MC724P 60p MJ420 80p MJ421 80p MJ2801 #1.10 MJ2901 £1.50	ZTX501 15p ZTX502 20p ZTX503 17p ZTX504 40p	£2.20 2N3716 £2.85 2N3771 £2.00 2N3772 £2.05 2N3773 £2.35 2N3791 £2.75 2N3819 35p 2N3820 55p
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BC114 BC115 BC116 BC116A BC117 BC118 BC119 BC134 BC135 BC136 BC137 BC138	20p 20p 20p 25p 20p 20p 30p 20p 20p 20p 20p 15p 20p 12p	#1.50 GM378A 55p MAT101 25p MAT121 25p MC724P 60p MJ420 80p MJ421 80p MJ2801 #1.10 MJ2901 £1.50	ZTX501 15p ZTX502 20p ZTX503 17p ZTX504 40p ZTX531 26p IN914 7p IN916 10p	\$2.20 2N3716 \$2.85 2N3771 \$2.00 2N3772 \$2.05 2N3773 \$2.35 2N3791 \$2.75 2N3819 35p 2N3820 55p 2N3823 50p 2N3823 50p 2N3824 95p
BC114 BC115 BC116 BC116A BC117 BC118 BC119 BC134 BC135 BC136 BC137 BC138	20p 20p 20p 25p 20p 20p 30p 20p 20p 20p 15p 20p 12p 12p 20p	GEAGE #21-50 GM378A 55p MAT101 25p MAT101 25p MAT121 25p MC724F 80p MJ420 80p MJ421 80p MJ2801 £1-10 MJ2901 £1-50 MJE340 50p MJE370 20p MJE371 80p MJE377 80p MJE377 75p	ZTX501 15p ZTX502 20p ZTX503 17p ZTX504 40p ZTX531 26p IN914 7p IN916 10p	22.20 2N3716 22.85 2N3771 52.00 2N3772 22.05 2N3773 52.05 2N3791 52.75 2N3819 35p 2N3820 55p 2N3823 50p 2N3824 95p 2N3903 20p 2N3906 25p
BC114 BC115 BC116 BC116A BC117 BC118 BC119 BC134 BC135 BC136 BC137 BC138 BC147 BC148 BC149 BC153 BC153	20p 20p 20p 25p 20p 20p 30p 20p 20p 15p 20p 12p 12p 20p 20p	GBX.05 GM378A.55p MATIO1 25p MATIO125p MATIO125p MOTICAL 60p MU420 80p MU421 80p MU421 80p MU421 80p MU421 80p MU421 80p MU4201 21-50 MU420 MU4201	ZTX501 15p ZTX502 20p ZTX503 17p ZTX504 40p ZTX531 26p IN914 7p IN916 10p IN4148 7p	22.20 2N 3716 22.85 2N 3771 22.00 2N 3772 22.37 22.37 22.37 22.35 2N 3791 22.35 2N 3829 2N 3820 2N 3820 2N 3822 2N 3823 2N 3824 2N 3824 2N 3829 2N 382
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HENRY'S LOW INTEGRATED CIRCUITS

BRAND NEW FULL SPECIFICATION TTL74 SERIES BRANDED FAIRCHILD, I.T.T. AND TEXAS

	BRANDED FAIRCHILD,	18 W BO	PARE	MINU DE	TOPO	_
	DEVICES MAY BE MIXED TO QUAL	TAX NO	K GOAM	25-99	TOOL	'250+
No.	Description	1-11	12-24		100+	
7400	Quadruple 2-Input NAND gates	20p	18p	14p 16p	14p 14p	12p
7401	Quad 2-input open collector NAND gates	20p	18p		140	12p
7402	Quad 2-input NOR gates	20p	18p	16p	14p	12p
7403	Quad 2-input open collector NAND gates	20p	18p	16p	14p	12p
7404	Hextuple inverters	20p		16p	14p	12p
7405	Hex inverters with open collector outputs	20p	18p	16p	14p	12p
7410	Triple 3-input NAND gates	20p	18p	16p	14p	12p
7413	Dual 4-input Schmitt triggers	30p	27p	25p	22p	20p
7420	Dual 4-input NAND gates	20p	18p	16p	14p	12p
7430	Single 8-input NAND gates	20p	18p	16p	14p	12p
7440	Dual 4-input NAND buffer gates	20p	18p	16p	14p	12p
7441	Dual 4-input Schmitt triggers Dual 4-input NAND gates Single 8-input NAND gates Dual 4-input NAND buffer gates BCD-Decimal decoder/Nixie driver	75p	72p	70p	60p	55p
7442	BCD-Decimal decoder (4-10-line) TTL O/P	75p	72p	70p	60p	55p
7443	Excess 3-Decimal decoder TTL outputs	£1 - 00	95p	90 p	80p	70p
7447	BCD-Decimal 7 seg, decoder/indicator driver	£1.75	£1 · 60	£1 · 45	£1.30	£1·15
7448	BCD-Decimal 7 seg. decoder/driver TTL O/P	21 - 75	21-60	21-45	£1.30	£1·15
7450	Expand dual 2-input AND-OR-INVERT gates	20p	18p	16p	14p	12p
7451	Dual 2-wide 2-input AND-OR-INVERT gates	20p	180	16p	14p	12p
7453	Quad 2-input expand AND-OR-INVERT gates	20p	18p	16p	14p	12p
7454	4-wide 2-input AND-OR-INVERT gates	20p	18p	16p	14p	12p
7460		20p	18p	16p	14p	12p
7470	Bingle J-K flip-flop (gated inputs)	30p	27p	25p	22p	20p
7472	Single J-K flip-flop (gated inputs)	30p	27p	25p	22p	20p
	Dual J-K flip flop	40p	37p	35p	33p	30p
7473	Duar J.K mp nop	40p	37p	35p	33p	80p
7474	Dual J-K flip flop	45p	42p	40p	38p	35p
7475	Quadruple bistable latch	40p	37p	34p	31p	28p
7476	Dual J-K filp-flops with Preset and Clear		72p	67p	59p	55p
7480	Gated Full Adder	80p	£1·15	£1.10	£1.00	90p
7481	16-bit read/write memory	21.25		70p	65p	60p
7482	2-bit binary Full Adder	87p	80p		80p	73p
7483	4-bit binary Pull Adder	£1.00	90p	85p		
7484	16-bit RAN with gated write inputs	90p	85p	80p	75p	71p
7486	Quadruple 2-input Exclusive OK gates	45p	41p	88p	35p	33p
7490	BCD decade counter	75p	70p	65p	60p	55p 70p
7491	8-bit shift register	#I - 00	95p	90p	80p	
7492	Divide twelve counter	75p	70p	65p	60p	55p
7493	4-bit binary counter	75p	70p	65p	60p	55p
7494	Dual entry 4-bit shift register	SOP	75p	70p	65p	60p
7495	4-bit up-down shift register	80p	75p	70p	65 p	60p
7496	5-blt parallel/serial in/out shift register	£1.00	97p -	95p	90p	83p
74100	BCD decade counter S-bit shift register Divide twelve counter 4-bit binary counter Dual entry 4-bit shift register 5-bit parallel/serial in/out shift register 5-bit parallel/serial in/out shift register 8-bit bistable latch Hextaple Set-Reset latches Monostable multivibrators BCD-Decimal decoder/Nixle driver BCD-Decimal decoder (1-4-line) TTL OP	£2.50	22.30	£2-00	21 - 75	£1.50
74118	Hextuple Set-Reset latches	£1 · 00	9 5 p	90p	80p	70p
74121	Monostable multivibrators	60p	55p	50p	46p	41p
74141	BCD-Decimal decoder/Nixie driver	£1-00	95p	90p	80p	70p
			£1-40	£1 · 30	#1·10	£1-00
74150	16-bit data selector/multiplexer	£3 · 35	£3-20	£2 · 95	22-15	22.05
75151	8-bit data selector/multiplexer	£1·10	95p	90p	80p	70p
74158	Dual 4-line to 1-line data selector multiplexer	£1 · 35	£1 · 27	£1 · 20	21-15	£1-10
74154	16-bit decoder/demultiplexer	£2·00	£1.75	£1 · 55	21-30	£1 · 05
74155	Dual 2-line to 4-line decoder/demultiplexer	£1 · 55	21 - 47	£1 · 35	21-10	£1·05
74156	Duai 2-line to 4-line decoder/demultiplexer	21-55	21 - 47	£1 · 35	21.10	21.05
74190	Sync decade up-down counter, 1-line mode	21 - 95	£1.85	21.75	£1 · 60	£1-50
74191	Sync 4-bit up down counter, 1-line mode	21-95	£1 · 85	21 - 75	£1 · 60	£1-50
74199	Sync decade up-down counter, 2-line mode	22.00	21.90		£1 · 65	£1 · 55
74193	Sync 4-bit up-down counter, 2-line mode		£1 · 90		41-60	\$1.40
74196	Asynchronous presettable decade counter	£1 - 50	£1-40	£1 · 30	£1 · 10	£1-00
74197	Asynchronous presettable 4-bit binary counter	£1 - 50	£1 · 40	£1.30	#1-10	£1 · 00
Comp	ete data on the above in booklet 20 payes, Ref. 29,	issue 2 a	15p post			
Texas	I.C. Handbook. Complete information on 10	0 types.	40p. Pos	t 10p.		
Inter	ated circuit sockets 14 pin D.I.L. 25p; 16 pin D.	L.L. 30p.				
B						

INTEGRATEO CIRCUITS MFC4000P 5 MFC4010P 6 55p 60p 42 50 21 50 1C12 PA246 TAD100 TAD110 MC724P 702C (TO5) 709C (TO5) 709C (D.I.L.) 723C(TO5) MC1303P MC1304P 81403D 1C12 £1.50 £1.50 50p 75p 45p 45p 45p 41.00 80p 22.00 22.25 £1.50 75p 40p 81.403D 741C(DIL) 914(TO5) 923(TO5) TOSHIBA 24-47 TOSHIBA

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25 100 500 1000 1- 25 24 + 400m/w BZY 88
series* 12p 10p 8p 7p 6p
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When I read about the Signal Injector by Alan Jardine I was reminded of the poor beginner.

Following the instruction to solder the leads direct on to the cell will result in heating up the electrolyte and a very short life for the cell.

Perhaps this is not important as the choice of a push-on/push-off switch allows no easy means of knowing if the thing is on or off. Very few beginners will remember to test each time, and cell life will be short it is expected. A push button perhaps?

The blind cannot be expected to lead the blind, and beginners are usually short of experience.

R. Quorn Sussex

Of your two points concerning the Signal Injector, the first is a bit exaggerated. It is true that the cell life will be reduced by applying heat (from soldering iron) to the battery terminals but this is only negligible for the time required to execute the connection.

To install a holder to suit this type of battery would increase the cost by about 40 per cent.

We agree that it will be difficult to tell if the unit is on or off when not in use, but it can be determined; when the unit is "on" the push button will feel "loose" but in the "off" position this looseness disappears.

If this proves unsatisfactory a push-to-make release off type

can be substituted.

More Accurate Timer

May I thank you for publishing another article combining the hobbies of electronics and photography (ref. Darkroom Timer, March issue).

Although of excellent design, I feel it must be stated that a timer with only a 5 second timing intervals is not nearly accurate enough for the demands of the high quality black and white or well balanced colour prints that are required. However, with a small modification, I have found that the timer may be converted to an accurate piece of equipment having a timing range of 5 to 45 seconds in one second steps.

The modification requires four extra components, which are a 5 position two-pole switch (S4), VR5, VR6 and VR7 which are

skeleton presets of the values, 5 kilohm, 10 kilohm and 20 kilohm respectively.

These components form an additional timing circuit which is connected in series with the

original (Rt).

Position 1 of the switch has no further resistance and acts as a short circuit; position 2 connects VR5 into circuit, whilst position 3 connects VR6; position 4 connects VR5 and VR6 and position 5 connects VR7 into circuit.

Each position of the new switch is to represent a further one

second delay.

Position 1 of course, has no further delay, position 2 however, will give a one second delay, position 3 two seconds etc. when the presets are set as they were in the original timing circuit.

Now, any time, in one second steps may be selected from 5 to 45 seconds by selecting the required 5 second range, plus the required extra time (if any) on the new switch.

D. G. Smith Emsworth, Hants.

Components

Let me say first of all how much I enjoy your magazine and as a newcomer to electronics I find your Teach-In articles very interesting and also Shop Talk, etc. However, I wonder if I may make a suggestion?

I constructed your Demo-Deck and find that in following this series for a month or so there is a list of the more minor components used in the experiments and I wondered if it would be at all possible, either, preferably, if you could publish the list of all the components that would be required for the rest of this series in one complete list or if possible broken up into the individual months during which they will be required.

The reason I say this is, that I,

like many of your other readers no doubt, have no local supplier of components in my immediate vicinity and it usually means a trip to Edinburgh or Glasgow to purchase these components.

However, if I could have a full list this would make things much easier for me. It would also make it much easier to send off a full list by post to a mail order firm rather than asking for two or three small components every month or so. I wonder if this could be done.

I am very grateful to you and wish you every success for your future publications.

> R. L. Grant Scotland

It was our intention to publish an advanced list and in future we shall be publishing, at the end of each Teach-In every month, a list of components additional to those you have already acquired.

Calling Gloucester

Now that I'm receiving your magazine on regular order and greatly enjoying it, I feel that I ought to go a stage further in order to get any lasting benefit from your guidance.

Can I please find out through your pages how many people in the Gloucester area are willing to ask for, and attend, an evening class on useful, basic "every-

day electronics"?

Should anyone be interested, could they please write to me at the address given, then provided enough wish it, our local Education Authority can be approached with evidence that the need for such a class does exist.

Many thanks for giving me a chance to ask for these people through your very sensible maga-

zine.

E. L. Payn 82 Innsworth Lane, Longlevens, Gloucester GL2 0DE

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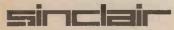
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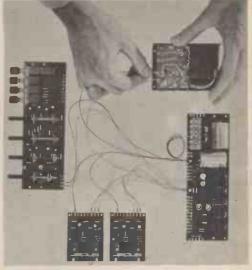
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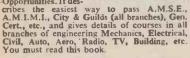
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