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- Component Packs - Most kits are available as separate packs (e.g. PCB component sets, hardware sets etc). Prices in our FREE catalogue.
- Ordering - Full ordering details, delivery service, and sales counter opening - outside back of this issue.



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# Sinclair XX81 Personal the heart of a system that grows with you. 

1980 saw a genuine breakthrough the Sinclair ZX80, world's first complete personal computer for under £100. Not surprisingly, over 50,000 were sold.

In March 1981, the Sinclair lead increased dramatically. For just $£ 69.95$ the Sinclair ZX81 offers even more advanced facilities at an even lower price. Initially, even we were surprised by the demand - over 50,000 in the first 3 months!

Today, the Sinclair ZX81 is the heart of a computer system. You can add 16 -times more memory with the ZX RAM pack. The ZX Printer offers an unbeatable combination of performance and price. And the ZX Software library is growing every day.

## Lower price: higher capability

With the ZX81, it's still very simple to teach yourself computing, but the ZX81 packs even greater working capability than the ZX80.

It uses the same micro-processor, but incorporates a new, more powerful 8 K BASIC ROM - the 'trained intelligence' of the computer. This chip works in decimals, handles logs and trig, allows you to plot graphs, and builds up animated displays

And the ZX81 incorporates other operation refinements - the facility to load and save named programs on cassette, for example, and to drive the new ZX Printer


Every $7 \times 81$ comes with manual - a complete course in BAS: first principles to complex prog


## Higher specification, lower price -

 how's it done?Quite simply, by design. The ZX80 reduced the chips in a working computer from 40 or so, to 21 . The ZX81 reduces the 21 to 4 !

The secret lies in a totally new master chip. Designed by Sinclair and custom-built in Britain, this unique chip replaces 18 chips from the $\mathrm{ZX} \times 0^{\prime}$

## New, improved specification

- Z80A micro-processor - new faster version of the famous Z80 chip, widely recognised as the best ever made.
- Unique 'one-touch' key word entry: the ZX81 eliminates a great deal of tiresome typing. Key words (RUN, LIST, PRINT, etc.) have their own single-key entry.
- Unique syntax-check and report codes identify programming errors immediately.
- Full range of mathematical and scientific functions accurate to eight decimal places.
- Graph-drawing and animateddisplay facilities.
- Multi-dimensional string and numerical arrays.
- Up to 26 FOR/NEXT loops. - Randomise function - useful for games as well as serious applications. - Cassette LOAD and SAVE with named programs.
- 1K-byte RAM expandable to 16K bytes with Sinclair RAM pack.
- Able to drive the new Sinclair printer.
- Advanced 4-chip design: ras... processor, ROM, RAM, plus máster chip-unique, custom-built chip replacing 18 ZXE ?



## Kit or built -it's up to you!

You'll be surprised how easy the ZX81 kit is to build: just four chips to assemble (plus, of course the other discrete components) - a few hours' work with a fine-tipped soldering iron. And you may already have a suitable mains adaptor -600 mA at 9 VDC nominal unregulated (supplied with built version)

Kit and built versions come complete with all leads to connect to your TV (colour or black and white) and cassette recorder.


## uter-




## NEWS:NEWS:NEWS:NEWS:NEWS:NEWS:NEWS

# DIGEST 

## 'Ears A Novelty

T-his new calculator has been specifically designed for Arab oil producers who wish to calculate up-to-the-minute profit figures while driving between oil wells. The device may be unobtrusively powered from the car battery via the discreet power lead, and for road safety one hand can always remain on the steering wheel since the buttons are operated by ear.

Well, that's what we first thought when this photograph fell out of our mail, but a glance at the accompanying press release revealed a more murdane explanation. Philips have received a share of an order from the Kingdom of Saudi Arabia for the delivery and installation of a national automatic mobile telephone network. Covering 32 cities and several of the nation's traffic corridors, the network involves $\mathbf{1 8 , 0 0 0}$ mobile telephones and 48 basestation sites, half of which are to be supplied by Philips. The full duplex system operates in the $\mathbf{4 2 0 - 4 7 0} \mathrm{MHz}$ band and provides 866 channels nationwide. The mobile telephone network can be automatically connected into the 'fixed' telephone system, allowing any number in the worid to be called. Facilities include a push-button dialling, dialled-number LCD display, automatic display/keyboard illumination to suit the lighting conditions, lastnumber recall and malicious-call tracing (wonder what the Saudi penalty for that is?). Naturally, the whole kaboodle is built round a microprocessor, and will give Saudi Arabia one of the most modern systems in the world.

Somehow our explanation seemed so much more fun.



## AIM <br> To Please

f you're a fan of the AIM 65 'bareboard' microcomputer you will doubtless be glad to know that you can now cover up its naked little body with an accessory enclosure from Rockwell. The ABS enclosure has a brown non-reflective crinkle finish, on-off and reset switches and internal AC lines. An optional power supply providing +5 V at 3 A and +24 V at 500 mA is also available with the enclosure. Available from all AIM 65 distributors as well as many computer store dealers, the enclosure has a refail price of about E90 with the PSU and $£ 50$ without.


## Long Live Batteries

And now, the battery that goes on Aworking after the battery that goes on working, stops. Israel's Tel Aviv University has developed a unique lithium-sulpher battery which yields one of the highest energy densities available. The cell can be used for a period of up to 10 years, yet is more compact and cheaper to construct than conventional long-term batteries. Existing in prototype form only at present, projected uses include microcomputer power failure back-up, calculators and watches, inaccessible instrumentation (eg military equipment) and pacemakers, where you don't really want to stick new batteries in every couple of months.

## Save It!

A re you stuck in a mental rut as Aregards traditional mechanical and electrical devices? Do you worry about the energy costs of your company? Kill two birds with one stone and visit a one day seminar on microcomputers and energy conservation. The seminar, to be held on Tuesday 20th April at the CEGB in London EC1, will explore what computers can and cannot do to save energy, from controlling electricity demand to optimising combustion in boilers. The seminar fee is $£ 100$ plus VAT per delegate and further details can be obtained from Miss Mallory Barker, Scientific and Technical Studies, Norwich House, 11-13 Norwich Street, London EC4A 1AB (Te). 01-242 2481).


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## NEWS:NEWS:NEWS:NEWS:NEWS:NEWS:NEWS

## Improving Memories

Not many people know this, but computer RAM chips have a built-in mechanism for producing errors - the casing of the IC. Trace elements of radioactive isotopes found in the device packaging materials give off alpha particle radiation, causing soft errors in both static and dynamic RAM chips. Two companies have tackled this problem using two very different approaches.

Dow Ecrning decided that shielding the silicon surface seemed the most logical approach, and developed a silicone rubber which is applied to the memory chips in the last step before hermetic sealing. One syringeful of the liquid silicone ( $\mathbf{3 0} \mathbf{~ c c )}$ can coat 2400 static 4 K RAMs and give at least one order of magnitude failure rate reduction. Hitachi, on the other hand, have im proved the reliability of their new plastic-packaged 64K dynamic RAMs using a combination of design features. These include higher cell charge storage capacities and higher signal read-out levels, resulting in a two orders of magnitude improvement over non-coated devices. Hitachi expect these chips to replace conventional ceramic 64K dynamic RAMs in the majority of applications.


## This Is A Recording

Atelephone answering machine which uses a microprocessor to combine simplicity of operation with versatility of use has been launched by Panasonic. The KX-T1524BE (who the hell dreams up these names?) is a twin cassette system using standard cassette tapes for both incoming and outgoing messages, eliminating the need for 'endless loop' or other specialist tapes. A variety of operating modes are available to cater for most user requirements and by using an individually coded pocket bleeper, you can retrieve messages by calling your own number from anywhere in the world. Skip and repeat functions are speeded up by using a message search system; a tone is added to the beginning and end of each message which is easily identified by the search system. The unit is available from dealers of Panasonic Business Equipment. Now if only British Telecom could do something about wrong numbers

## OOPS

Acouple of boobs to comment on this month. First, the Music Processor (November '81). There are two R38s on the overlay; the one connected to pin 7 of IC5 is an error, but makes little difference as pin 7 is not connected internally. The published PCB has IC6 pins 5 and 6 connected together; pin 6 should only connect to C 15 so cut the extra track. Boards supplied by our PCB Service do not have this error. The lead from R51 is marked LED1 anode on the overlay; it should be LED3 anode. The point marked LED3 cathode should also go to LED1 cathode.

Second, two errors appeared in the Guitar Tuner circuit diagram (January '82). The supply pin numbering for IC3 is reversed; ( + ve is pin 4, -ve is pin 11), and IC3c is drawn incorrectly. Pin 5 goes to the V/2 rail only; R12 and C10 should go to pin 6. The overlay is correct. Incidentally, it appears that some people have been buying TMS $\mathbf{1 0 0 0}$ ICs from sources other than Magenta and finding that the project doesn't work. This is because the IC is a mask-programmable chip and different versions do different things (some of them are doorbell chips, for example). Only chips described as TMS 1000 MP0121 will work in this circuit and only Magenta sells them.

This is the last time that OOPS will appear in the Digest pages. From now on amendments and improvements to our projects will appear in the READ/WRITE pages.


## Big, Big, Bar

ED bargraphs are considerably more robust and visually attractive than -mechanical meters, but suffer from poor resolution due to the limited number of LEDs available. Until now, that is. This massive beast from the Hewlett-Packard stable contains 101 - count them - LED elements in a 10.58 cm long package; the wide elements ( 1.52 mm ) and $1 \%$ resolution result in accurate and reliable meter indication. The LEDs are red and configured as a common cathode arrangement to simplify interfacing; HP say the HDSP-8820 is particularly well-suited to industrial process control systems. In one off quantities the device costs E35.10 from HP authorised distributors for more information contact Enquiries Section, Hewlett-Packard Ltd, King Street Lane, Winnersh, Wokingham, Berks RG11 5AR.


## Just Dropped In

jew equipment approved by the Electricity Council for high voltage live-line testing of overhead power cables and switchgear busbars will incorporate taut-band meter movements from Sifam Ltd of

Torquay. One of the tests involved dropping the meters five times from a height of two metres onto a con crete floor; since Sifam won the contract we assume the photograph shows the meters being assembled prior to testing, and not repaired afterwards.


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## Mao-Tse Tongue

Chinese publishers and businesses can now handle text with the same electronic ease as their Western counterparts; Ferranti Computer Systems have just developed a unique Chinese word processing system. Text is input onto disc using Chinese phonetics, the computer's software helping the operator to select the correct character from a dictionary of over 8,000 . Text can then be edited, formatted into tables or forms, coded for telex transmission or printed on paper. Incoming telex messages can also be decoded. No technical knowledge is required to run the processor, just the ability to speak and read Chinese. Bet Mao would have liked one for his little red


## Is VIC There?

Whenever a major company launches a peculiarly limited computer, someone somewhere will start to design improvements for it. It happened to the ZX81 and now B \& B Computers of Bolton have produced an add-on unit for Commodore's VIC-20. The 'black box' will increase the $\mathbf{2 2}$ column display to $\mathbf{4 0}$ columns and expand the memory from the meagre 3 K to 35 K ; all connecting cables and an additional PSU are in cluded. The expansion unit costs £220 plus VAT by mail order from Beelines Ltd, Freepost, Bolton BL3 6YS; the comparable CBM system would cost $£ 337.81$ plus VAT.

## Tangerine Toolkit

Lard on the heels of their high Idefinition Programmable Graphic Module, the Tangerine User Group are now offering a powerful Toolkit for comprehensive control of both PGM and chunky graphics on the Microtan 65. As well as fast plotting from both machine code and BASIC instructions, a number of extras such as BASIC line renumbering are offered. The PGM Toolkit is supplied in a 2716 EPROM with a comprehensive user manual and costs $£ 26.69$ for non-members.


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## ETI MOBILE

With the publication next month of the digital PWM board (held over for lack of space) and the infra-red proximity detector, our mobile robot is starting to take shape. Even if you're not a robot freak, the designs are versatile enough to be used in a whole range of other applications. Something for everyone - that's the June ETI.

## BREADBOARDING SYSTEMS

As we pointed out last month, there are many new breadboarding systems being released. Unfortunately most of them are so new that they didn't arrive in time for us to write the feature this month. Never mind, though - it's all here now and ready for our in-depth testing and review. If you're not keen on etching PCBs, wait with bated breath for the revelations that will unfold before your very eyes in next month's issue.

## NEGATIVE ION GENERATOR

One or two letters have come flooding in, demanding an ion generator design. Ever eager to please, we'll be publishing a simple experimental device next month so that you can see for yourself whether the claims are justified. Cleaner air, fewer headaches, allergies alleviated, a happy disposition even while reading your morning paper; all this can be yours at no risk of electrocution (we've been quite careful about that). Delve into bioelectronics with ETI next month.

## SERIES 5000 MOSFET AMP

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[^1]

# COLUMN LOUDSPEAKER DESIGN 

# The column loudspeaker is an example of a directional sound source which is of special use if acoustic feedback is a problem; it is most commonly found in sound reinforcement or public address applications. David Hornsby describes a novel design that can be made at a fraction of the cost of its commercial equivalent. 

Astandard loudspeaker tends to radiate sound in all directions, both forwards and backwards. It is helpful to look at this sort of response on a polar diagram (Fig. 1); the circle round the sound source shows that the sound loudness is about the same in all directions. If the loudspeaker is now placed in an enclosure then sound is allowed to radiate forwards only and we have a 'unidirectional' source. The polar diagram for it in Fig. 2 shows a balloon-like shape for the sound radiation pattern, which now covers an angle of slightly less than $180^{\circ}$. The dotted line shows the response if the enclosure lets a little sound out backwards.

If polar diagrams are new to you, these two diagrams will probably have given you a fairly good feel for what they are all about. They are similar to the contour lines on a map, but instead of showing height they show the sound intensity or loudness. The further the line on the polar diagram is from the sound source at the centre then the louder the sound is in that direction.

## One-Way Sound

The unidirectional sound source is the one most of us use in our homes and cars but it's not very good for live performances where the microphone(s) is in the same area. Diffraction effects, echoes and reverberations all help to spread the sound back from the loudspeaker to the microphone so that as soon as the sound is turned up, positive feedback makes the system oscillate and howl.

There are one or two different solutions to this problem but the most common and probably the best is to use a highly directional sound source. This tends to concentrate the sound into the area where it is needed, the audience, but well away from the microphone.

One type is the horn loudspeaker which has good directional properties and is also very efficient electrically. Unfortunately it has to be physically large to be effective at low frequencies. The fog horn at your local lighthouse (you do have one, don't you) and the PA in a cinema have space for large horn


Fig. 1 Polar diagram of a sound source which radiates equally in all directions (approximated by an unmounted loudspeaker). The sound source is at the centre of the diagram.


Fig. 2 Polar diagram of a sound source which radiates mostly forwards (approximated by a loudspeaker mounted on an infinite baffle enclosure). If a little sound is allowed out backwards the dotted line applies.


Fig. 3 Polar diagram of a column loudspeaker at $\mathbf{1} \mathbf{~ k H z}$.
units, but in the domestic scene we either have to use a folded horn design or just use tiny horns for high frequency tweeter applications.

The other directional type of loudspeaker in common use is the 'line source' or column loudspeaker, and that's the one we're going for here. The theory tells us that all we need is a loudspeaker with a cone which is long and narrow, rather like an elliptical loudspeaker taken to the extreme. Put that in a similarly long and narrow enclosure and that should be it! We do still have the problem that unless we allow the length to be at least a few feet then we will lose the beaming effect on the low frequencies, but there's a far worse problem - how do we actually get hold of our crazily-shaped loudspeaker? Does such a beast even exist? Actually it probably could be made with an electrostatic speaker but that's not for us. Instead we can approximate a line source with several conventional round speakers stacked in a line. Commercial designs use three or more, often quite a few more, and this works well. Our design uses five speakers spaced evenly along an enclosure of one metre length.

## Directional Characteristics

It is at this stage that we must look again at a polar diagram for our design, Fig. 3, and this is where this diagram begins to give us some useful information. The first thing to note is how the shape is drastically changed from a balloon to a series of fingers of various sizes. The largest finger is the main beam or lobe of our column loudspeaker while the smaller fingers are unwanted side lobes. If you have seen interference patterns on a ripple tank then you will probably understand the reason for this sort of pattern. To improve the directional properties of the speaker system still further, we want to reduce the side lobes and enlarge the main lobe. It would probably also be useful to have a slightly broader main lobe, since it is unlikely that we can

Fig. 4 Polar diagram of a column loudspeaker at 1 kHz with graded aperture.
arrange for the audience to be confined into too narrow a region.

Without going into all the maths of the solution, both these aims may be reached by a process known as 'grading' or 'tapering' the aperture. This is a little trick that is used in all sorts of situations, not just column loudspeakers. Microwave dish aerial systems often do just the same, for example. In our case, tapering the aperture simply means that we must arrange to evenly decrease the power fed to each of the individual loudspeakers as we move away from the central one on the column. The effect of this is shown in Fig. 4. Note that these diagrams both apply only at one frequency, 1000 Hz . At higher frequencies the lobes are narrower and more numerous, but they become wider and less numerous at lower frequencies until below about 500 Hz , the wavelength of sound is comparable with the length of the column and the beaming effect begins to fail. Fortunately acoustic feedback is likely to be worst at frequencies well above 500 Hz so one metre is as long as we need to make the column.

You may occasionally come across giant column loudspeakers which are also curved so that they look concave from the listener's vantage point. This produces the same effect as tapering but is not necessary except for systems much longer than one metre

## Electrical Design

So now a way of arranging the power feed to each speaker has to be devised. If series resistors of appropriate values are wired in with the speakers, then, although things work well enough acoustically and electrically, we will have an inefficient design which wastes much of the power of the amplifier as heat in these resistors. The common commercial solution is to forget about tapering altogether, or for expensive units to use a special matching transformer with tappings for each individual
speaker. This not only adds to the cost but also to the weight of the final product. Don't forget that no transformer has yet been designed which gives zero distortion, so that's yet another problem. While pondering this (in the bath - where else!) the author devised what seems to be a splendid engineering solution; that is, one that cheats the situation by winning several points at one go but without making any serious concessions. The key is to use identical speaker units but with different coil impedances. After many calculations with a range of different combinations, one design stood out as being almost ideal. It produces an effective impedance of 6.15 ohms, gives an even tapering and uses just 8 ohm and 15 ohm speaker units which are readily available.


Fig. 5 The wiring diagram for the 8 and 15 ohm speakers. Note how each speaker is wired in phase. For 10 W speaker units the power handling is about 25 W (actually 24.375 W ). At right is shown the profile of the graded aperture.

The electrical set-up. is given in Fig. 5 and for nominal 10 W units produces a speaker system of 25 W capability. The actual make of loudspeaker unit doesn't really matter provided you can get both 8 and 15 ohm units in the same style. The original design used R.S. Component's wide-range six-and-a-half inch loudspeakers which have given reliable service for over four years now. Some may object that five 10 W speakers ought to give a system capable of more than 25 W . It is, of course, the tapering of the system which causes this reduced power rating,



Fig. 6 An alternative way to wire the speakers. This gives a power handling of nearly 40 W but is not recommended as the aperture is not correctly graded. The profile of the aperture is irregular, as shown.
but its electrical efficiency is fair and there is no real problem. It is in fact possible to rewire the individual units so as to increase the power rating to 40 W , as shown in Fig. 6, but the tapering goes out of the window with this arrangement and it is not recommended.

Calculations show that the series/parallel combination of speaker units in our design gives an effective impedance of 6.15 ohms. This is just about ideal and suits the 4 to 8 ohm range that most power amplifiers are designed to feed. If you happen to have one which cannot drive impedances less than 8 ohms then you will need to add a 2 ohm series resistor to get things right. However, most column loudspeakers are necessarily mounted some distance from the amplifiers and the leads' resistance may provide some or all of this extra 2 ohms if you are lucky.

## A Case In Point

The cabinet for the design may be made from chipboard. Three-quarter-inch thick is about the right grade for this job. If you are going to use the R.S. speaker units then, provided your woodworking skills are fair, it is only necessary to refer to Fig. 7 for all the details. If you have or can gain access to a circular saw (what about woodwork evening classes?) the task is that much easier. None of the dimensions are that critical, but the overall volume has been designed to match the suspension characteristics of the speakers themselves and should be kept the same. The unusual cross-sectional shape is not an essential


Fig. 7 Cabinet details - all dimensions are in millimetres. The total cabinet volume is 0.324 cubic metres. A suitable material to use is $3 / 4^{\prime \prime}$ chipboard and this thickness needs to be added to the dimensions shown where appropriate.
part of the design either but was chosen so that the column could be neatly and permanently mounted on a wall and still point in the right direction. If your intended use is stage work then a square or rectangular cross-section giving the same volume would be easier to make.

Take care to close all joints with enough glue to make the unit reasonably airtight since this is a requirement of this type of speaker unit's cone suspension. The inside of the cabinet is filled with acoustic wool or similar sound-absorbing material so as to reduce internal sound reflections which otherwise give an unnatural colouration to the performance. I once knew a musician who insisted that internal lining of an enclosure reduced the high frequency response, but he had simply come to enjoy a particular type of distortion - don't leave it out! The best way to fix it is to tack it on lightly before the front is put on the enclosure. If it is not fixed it will soon fall to the bottom and lose most of its effect; if it is glued it tends to become compressed on to the glue which again cuts down on its absorption properties. Similarly, use a proprietary make of grille material for the front rather than any old material or again you will distort the sound. Most probably it will be the high frequencies that you lose this time if you are tempted to use the spare curtains because they are the right colour!

If the final unit is to be attached to a wall, a small screw recess can be provided near the top of one side for this. Most ironmongers stock screw-on brass plates that are ideal as a reinforcement for this. Don't forget to provide electrical connections on the back before the unit is assembled. Suitable types are available from the same sources as supply the acoustic wool and grille material (and the speakers themselves for that matter). The finish on the outside of the cabinet is obviously a matter of personal choice. If you wish to make a feature of it you can use a wood veneer or vinyl covering to achieve a smart appearance. The original design was made to appear unobtrusive (if that's not a contradiction in terms) by simply painting it the same colour as the wall it was to hang on, and this worked very well.

## Performance

In assessing how well the design works we must first decide what it is we are looking for. With a speaker system intended for hi-fi applications we might look at the frequency response and phase linearity, for instance, but this design is for sound reinforcement purposes. The chief needs are to reduce acoustic feedback by efficient beaming of the sound and to improve the audibility of whatever is behind the microphone. The design was originally made to meet the needs of a church of moderate size (about 50 by 30 feet) for both music and singing from the music group at the front, and for speech from the pulpit(but not both at the same time!).

Judged by these standards the final product is totally effective; the beaming effect is very noticeable. When it was tested in the living room at home before installation the sound appeared to be thrown forwards towards the listener in a way that the conventional speaker cannot achieve. With a pair of speakers now hanging on side walls each side at the front of the church, their base being six feet from the floor and the axis of each speaker pointing towards the floor at the mid point of the back wall, the comparison with the old temporary single speaker units is really quite spectacular. At the front of the church the sound is beamed high over people's heads and so is not deafeningly loud. At the back, however, the beams reach down to ear level and the sound seems every bit as loud as at the front even though you are further from the speakers. What is more, the sound, particularly speech, is strangely clearer. The effect is perhaps not unlike that in the Whispering Gallery in St. Paul's Cathedral, where you might be surrounded by background noise yet can hear a whisper with startling clarity from a spotexactly opposite. The speakers do not whisper but the sound seems to surround you in the same way.


Fig. 8 The working layout of the speakers in the church.

The power handling of the column loudspeaker, 25 W , is more than adequate. The two units are driven by mere 15 W amplifiers but even these are never turned up anywhere near full volume. Acoustic feedback is no longer a critical problem, no mean achievement in a stone church building. The improvement is so pronounced that the music group now need to be provided with extra speakers to provide foldback.

The frequency response of the units is essentially that of the individual loudspeakers - about 70 to 16 kHz for the R.S. units in an enclosure of this volume. Purists will notice and object to the lack of the extreme high frequency element, but this is unimportant in this application. In fact a design of this type will have rather better characteristics than the straight theory predicts since the matching of the system to the air is improved with the larger surface area of many loudspeaker cones. At low frequencies in particular it appears the response goes down well below 70 Hz although no measuring equipment was available to make quantitive measurements. If operation above 16 kHz is important for you don't despair: add a horn tweeter and mount it on top of the cabinet.

The overall impression of the speakers is of clean effortless performance, lacking only in that extreme high frequency content. They have been used regularly for four years now with $100 \%$ reliability. Applications have included not only the live sound sources mentioned earlier but also the playing of taped music and use for film shows. Once when playing back a taperecorded voluntary from the pipe organ, several members of the congregation admitted to me afterwards that they had to look at the organ to check that it was not live playing - quite remarkable really when you think that the organ is at the back of the church and the column speakers at the front! This is the result obtained with directional sound: it seems to come directly to you. If you want to test for yourself and are in range of North Buckinghamshire, why not pop into Holy Trinity Church, Deanshanger and make up your own mind?

One last note of caution for you: do position your column speaker the right way round, that is vertically. You will have seen from Figs. 3 and 4 how the sound beam spreads out from the system. Possibly because this spreading is the opposite of what might at first be expected or perhaps because of plain ignorance, column loudspeakers are occasionally positioned the wrong way round! In fact I know of one not many miles from my home, where, in a specially converted stable, a column speaker is attached horizontally to an old oak ceiling beam. Wild horses wouldn't drag the exact location of the stable from me (pun intended - groan); I enjoy the little theatre too much to want to upset them.

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# ECONOMICAL HEATER CONTROLLER 

## Save energy and master the meter; control your power consumption with the ETI Wattmiser. Design and development by Phil Walker.



After this winter's snow, frost and electricity bills we are probably looking for ways to save a little money without losing comfort or convenience. This project is designed so that appliances such as electric fires or water heaters can be set to turn on a little before they are needed. This will ensure that the room or tank of water, etc is comfortable or ready for immediate use when required, but has not been consuming those precious units all the time.

The device could also be used to control other things such as tape recorders, radios and so on where the precise timing is not too critical.

## Operation

The device operates by dividing the mains frequency ( 50 Hz ) by 180,224 to get a signal with a period of
about 1 hour. The actual division required was 180,000 but the 180,224 was more easily achieved and the error involved was less than 1 part in 500 , or about 40 seconds in the maximum period of 9 hours delay.

This 1 cycle per hour signal advances a divide-by-10 counter which has 10 separate outputs. Each output is active for 1 hour and only one is active at a time. One of the outputs is selected to drive the mains power switching triac - thus giving an 'on' period of 1 hour after a selectable off period. Also it is possible (if SW4 is on) to have the output come on permanently after its set delay (this is always the case on the 9 hour delay setting).

To use the device, set the number of hours delay with SW3, select whether ' 1 hour only' or 'permanently on' operation is required with SW4 and
press SW2 to start the device working.

## Installation

When using the device, the mains supply must be fed from a plug or switched fuse outlet containing a fuse rated for the appliance used. Also the wire used should be suitably rated round flex rated at about 15 A usually sold for electric fires or similar applications could be used in most cases.

## Construction

The construction of the PCB should present no great difficulties. The first thing to do is to ensure that the mounting holes are of a suitable size and that the correct ones are being used if the specified box is utilised. Next put some M3×12 mm bolts
through the mains power connection pads, heads on the copper side. Run some solder around these heads to secure them and improve the contact. Now fit the wire links and the other components with the exception of the switches, LED1 and SCR1. Make sure that D1 and D2 are the right way round. Also note that IC1 is the opposite way round to IC2, 3 and 4.

Next fit SCR1 and its heatsink to the board, using an $M 3 \times 6 \mathrm{~mm}$ bolt.

Use heatsink compound under SCR1 to improve thermal conductivity. Bend the leads out from SCR1 to go through the proper holes. Note that if the lower rating triac is used, the fixing hole nearer the contact pads should be used and vice versa. SW3 should now be fitted. If the contacts are on a 22 mm diameter circle they may be connected straight to the board, otherwise short lengths of stiff wire should be used. (If all else fails you could fix the switch to
the cover of the box and connect to the board using thin flexible wire leave enough to remove the lid when fitted).

Attach some lengths of flex ( $6^{\prime \prime}$ ) to the PCB at the remaining switch positions (SW2 and SW4), LED1 position and transformer connections. Fit the transformer diagonally into the corner of the box (the end with the wider spaced fixing bosses) and then the cable glands, neon, mains switch



Fig. 2 Pulse shaper action.

Fig. 1 Block diagram of the unit.

PARTS LIST

| Resistors (all 1 | 1/4 W, 5\%) |
| :---: | :---: |
| R1 | 15k |
| R2 | 220k |
| R3,4 | 100k |
| R5 | 27k |
| R6 | 1k2 |
| R7 | 820 R |
| R8 | 56R |
| R9 | 100R |
| Capacitors |  |
| C1 | 1000u 40 V axial electrolytic |
| C2 | 1 no ceramic |
| C3 | 14063 V axial electrolytic |
| Semiconductors |  |
| IC1 | 4093 |
| IC2 | 4060 |
| IC3 | 4516 |
| IC4 | 4017 |
| IC5 | MOC3020 |
| Q1 | BC477 |
| D1,2 | 1N4148 |
| BR1 | W01 bridge rectifier ( 100 V , 1 A) |
| SCR1 | TIC263D ( $400 \mathrm{~V}, 25 \mathrm{~A}$ ) or TIC246D (400 V, 16 A) |
| LED1 | TIC246D ( $400 \mathrm{~V}, 16 \mathrm{~A}$ ) <br> 3 mm red LED |
| Miscellaneous |  |
| SW1 | DPDT mains on/off rocker switch (miniature) |
| SW2 | 1 pole push-to-make switch, momentary action |
| SW3 | 1 pole 12 way rotary wafer switch, contacts on $\mathbf{2 2 ~ m m}$ diameter circle |
| SW4 | SPST toggle switch |
| FS1 | 1A anti-surge fuse and panel fuse holder ( $\mathbf{2 0} \mathbf{~ m m}$ ) |
| LP1 | Mains neon indicator with integral resistor (panel mounting) |
| $\mathrm{r} 1$ | 6 VA mains transformer (9 V secondary, 240 V primary) |
| PCB (see Buylines); case (see Buylines); 2 off cable glands ( $\mathbf{8 - 1 0} \mathbf{~ m m}$ - see Buylines); 3 off 16 pin and one off 14 pin IC sockets (if used); heatsink (TO-220 style) knob; M3 nuts and bolts, spacers, etc. |  |
|  |  |



## BUYLINES

Most of the components used in this project are standard off-the-shelf items. The MOC3020, TIC263D and IIC246D may be a little harder to track down and can be obtained from TK Electronics. The case we
used was a BOC 450 from West Hyde Developments; the same company can provide the cable glands we specify. The PCB can be obtained from our PCB Service using the advert/order form on page 45.

## PROJECT : Heater Control

and fuseholder into the end-plate. Wire up as shown on the circuit diagram.

Finally fit SW2, SW4 and LED1 to the lid of the box and cut a hole through to take SW3 spindle. Wire up as per the circuit diagram. Mount the PCB on pillars using M3 bolts at such a height that the SW3 spindle protrudes far enough to take the knob when the case top is in place.

In use it is advisable to solder eyelets to the mains wires to fit on to the PCB bolts; alternatively the ends of the wires should be soldered to stiffen them.
NOTE: As mains voltages are present, care must be taken that no unearthed metalparts are accessible from the outside and that clearances between live and earthed parts are maintained under all circumstances.

## HOW IT WORKS

IC1a takes the raw AC 50 Hz signal applied to the bridge rectifer and converts it to a train of pulses at the same frequency. This is divided by $\mathbf{2}^{14}(16,384)$ by IC2. Further division is done by IC 3 which is connected with IC1c to divide by 11 (10).

This is accomplished by loading $\mathbf{1 0 1 1}_{\mathbf{2}}$ into IC3 each time it counts down to 0000 (2) At the $Q_{4}$ output of IC3 there is a signal which has a period of $20 \mathrm{mS} \times 16,384 \times$ $11=3,604,4 \mathrm{BO} \mathrm{mS}=3,604.48$ seconds. This signal is applied to the clock input of IC4, a divide-by- 10 device with 10 decoded outputs. The output of IC4 selected will change every $3,604.48$ seconds (about 1 hour) until the ' 9 ' output (pin 11) is high. At this time further clocking of IC4 is prevented by a signal at pin 13 via D1.

SW3 selects which output of IC4 drives the output power switch. This varies from permanently on (position 1), through zero
delay (position 2), up to 9 hours delay (position 11), to permanently off (position 12).

If SW4 is open then in the 0 to 8 hour delay positions the output will be on for 1 hour after the set delay only, whereas if SW4 is closed or a 9 hour delay is selected the output will stay permanently on after the delay period.

IC1c takes the output signal from SW3 and drives Q1 and thus IC5. IC5 is an optically isolated triac which enables us to have the logic circuitry safely at earth potential while switching the mains power with SCR1.

The last part of the circuit is that around IC1a and IC1d. This is the reset circuitry which ensures that the time period will be consistent and start when SW2 is released. IC2 and IC4 are reset to 0 by IC1a, while IC3 is set to 1011 (decimal 11) by IC1d.


With the lid off, the general layout of the project should be clear. With the large currents involved, the mains wiring is secured to the PCB with nuts and bolts.


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$16-32 \mathrm{~K}$...This area can be used for basic programmes and assembly language routines.
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# AUDIOPHIIE 

It's dents-in-the-desk time, as Ron Harris listens to some hernia-inducing Hitachi hi-fi. This MOSFET combo is mighty in more ways than one.


This month an excursion into MOSFET territory, made in the sure company of Hitachi - who can be regarded as something of pioneers in the field. Actually, this review comes about, in its present form, by the strangest of circumstances. Firstly, I was assembling a 'group-test' of four or five high-quality amplifier combinations of various technical bents. I know that sounds like an orgy in an all-male laboratory, but I can assure you my motives were pure.

As with all things in life, some arrive ahead of others, and the Hitachi came weeks ahead of the rest. (The Bishop's actress would have been proud of them . . .) In between times, there was a press launch from a major manufacturer for one of those few new models which can raise the eyebrows and clean out the ears in anticipation.

More detail than that I dare not reveal - the BOYS are leaning over my shoulder, hands on violin cases, even as I write this death comes but once and that painfully. Embargo dates shalt thou keep, lest thy tender bits be removed and cast upon the waters. Watch next month's Audiophile. If I'm still in possession of my facilities, all questions will be answered. (I'm too young to die, I still haven't had that romantic dinner with Felicity Kendal...)

Anyway, returning to the track upon which we embarked the fact that I had to tie up large chunks of next month, allied to the somewhat tardy arrival of all the other amplifiers, left Hitachi home-free this month with their HMA-7500 II power-amp and HCA-7500 preamp combination - unusual for its employment of MOSFETs.

This is a subject upon which we have received much correspondence - mainly requests to produce a project. Well we did - two years ago. Ahead of our time again, you see. Designs age, however, and that Ambit unit is somewhat irrelevant by today's standards of quality. If we were going to retain our reputation for state-of-the-art electronics, a complete re-design was called for. Next month you will be presented with the results of our efforts. A hundred and odd watts of pure magic. MOSFET magic.

## Hitachi 7500 Ils

The 7500 Ils are actually cheaper than the Mark Is, but offer more facilities for lesser amounts of green stuff. The preamp is well-decked out with variable turnover tone controls, two tape inputs, moving-coil pre-preamp and variable input parameters on moving magnet. In addition the volume control is bedecked with little decibel numbers and there is a loudness control, a mute switch and a subsonic filter.

The power amp is bloody huge. It would be polite to call it 'large', but polite here is insufficient. Bloody huge somehow captures the spirit of the thing in a way that 'large' simply fails to do. The front panel sports huge meters, too - carefully graduated in watts RMS into four and eight ohms. In addition there is a 'BTL' facility which allows the 80 W stereo to be transformed into 150 W mono at no sacrifice in quality. Two pairs of speakers are catered for, although some care in choice is required.

All inputs and outputs on the preamp are phono - the dreaded DIN is nowhere in sight. Three cheers for Oriental partisanship. Sometimes anyway.

## HCA-7500 II Preamp

At $£ 139$ this design represents a large number of buttons per pound. There are a restricted number of separate preamps available on the shop-shelves today and most cost considerably in excess of $£ 139$. Strange as it may seem to those descendents of Ebenezer Scrooge reading this, the Hitachi can be safely regarded as a BUDCET unit.

These days I am becoming increasingly suspicious of the disc-amplifier stages in all designs, especially if they are intended to cope with the moving-coil variety of groove followers.

Accordingly all preamps get somewhat of a beating about the equalisation if they pass this way. Well, it makes a starting point and we all need one of those, do we not? In addition the continual march of digital and high-quality 'super-discs' places ever increasing demands upon these stages. They are thus increasingly under the microscope.

Regrettably the 7500 did not pass this particular line unscathed. Technically it fared well under test. You can read the results for yourselves. However, in use, the pickup stage sounded below par - especially when compared to the rest of the 7500 set-up.

The moving-magnet amp added a certain warmth to the sound, which spoke of low-frequency colouration and clouded detail right up into the mid-range. The moving-coil eption added some noise but did nothing to cleanup the sound

These are absolute judgements you realise, relative to a reference far more expensive than the 7500 . Taking price into consideration alters the balance to the extent that the Hitachi is a good value performer for its cost and one which does not seem illmannered in all but the most exalted company.

## Pickup Downs

Apart from this, the HCA is a fine unit indeed. It is quiet and efficient and does its job with a minimum of fuss. Pulse testing the unit reveals a good handling of transients, a function of the unit's wide bandwidth and fast slewing, and conventional tests point to low-noise and flat frequency response.

Mechanically the controls operate smoothly and give the impression of having been well engineered. The volume control is a 'click-stop' type of high quality. Somehow, though, I found it awkward to use. It seemed to be wired the wrong way around! An odd observation I know, but if you tried it I think you would see what I mean. The dB graduations, which get lower in value as the volume goes up, don't help either, correct though they may be!

Cartridges tried with the HCA-7500 included the Dynavector Ruby and the DV20All, the Coldring C900ICC and the Shure MV30H reported upon last month. Tuner source was the trusty (and unsurpassed?) Pioneer TX9500.

Tape facilities allow for two decks, with monitor on either and dubbing between the two in either direction. There is an unobtrusive little push-button labelled "Source Copy" which enables the outputs to tape. Forget to push this and you can have all sorts of fun trying to find out "where has all the signal gone.

Three impedances are available for moving magnets ( 22 k , 47 k and 100 k ) and two input levels for moving-coil, ( 0.1 mV and 2.5 mV ) at 100 R input impedance. More useful perhaps would be capacitance switching for the MM units, to better control high frequency behaviour.

## Loudness Played Softly

I've never understood why a button which corrects for hearing losses at low levels is called loudness. Still I suppose'softness' conjures up images of fabric conditioners and woolly sounds.

If you must have one of these ill-named tone controls, then this is about the best you'll find! Most tend to wildly overdo things in the bass with an excess of 10 dB boost at 100 Hz and below - and no topend correction at all. Figure 1 shows the


Fletcher-Munson curves upon which all such correction is based and you can see the required topend lift quite clearly.

Because human hearing is less sensitive to extremes in the audio range, at low listening levels your perception of bass and treble is severely limited. Loudness controls are intended to compensate by shaping the frequency response into an inverse plot of the hearing sensitivity curve you see in the diagram.

At high levels, where frequency has far less effect upon perceived level, they should be inoperative.

As I said, the Hitachi HCA 7500 II boasts a somewhat more believable compromise than most.

## Powerful Amps

The HMA-7500 II is a hefty piece of work, sporting two wellendowed power supplies and those massive meters. It is possible to switch the machine into bridged mode, doubling the power available into eight ohms. Those PSÚs boded well for the burst power capability of the amp, an idea which seemed to be borne out by early listening tests - only to cause confusion in the lab!

The rated 75 W was delivered uncomplainingly from 5 Hz to $70 \mathrm{kHz}(\mathrm{RMS})$ and at 1 kHz 90 W was available. This with less than $0.01 \%$ total distortions (THD and IM). Burst power delivery proved to be a disappointing 104 W into 8 R - I would have expected greater things in view of the sheer size of the power supplies. Into 4R, delivery was, as near as dammit, the same 100 W . Into 2R strange things began to happen, courtesy of the protection, and no readings were feasible.

Lesson? Be careful with speaker choice if you intend to push the 7500 II near its limits.

## Burst Testing

I've had some puzzled letters - and one very aggrieved epistle for which 'tirade' is a good word - enquiring as to how I


Fig. 1 Above: the Fletcher-Munson graphs of equal loudness. The curves plot the amount of energy required to produce a sound of equal intensity at the ear, for varying frequencies.

Left: the inside story on the $\mathbf{7 5 0 0}$ preamp. Note the use of mechanical couplers to the input switching, which is PCB mounted near the input sockets. Just about everything that can be done to reduce noise, has been done.
obtain my oft-quoted burst delivery power measurements.
The basic idea is to better simulate what the amp is capable of doing with music, as opposed to sine or square waves, thus correlating more closely lab test results with listening tests. As a music signal is composed of a whole spectrum of short-duration tone-bursts scattered from $20 \mathrm{~Hz}-20 \mathrm{kHz}$ fairly randomly (??), I don't think that a sine wave at 1 kHz is a close enough approximation, even in short pulses.

My own way of obtaining a number is to gate a $1 \mathrm{mS}-10 \mathrm{mS}$ burst of white noise, modulated by a sine wave, into the unit under test and read the resulting power delivery into an eightohm load. At present I'm working on improving the load, so that it will better approximate a speaker.

The variable length of the input pulse is required to discover when and where the protection, or limiting, operates in the amplifier circuitry. A perfect amplifier would delivery $X$ watts into $8 R, 2 X$ watts in $4 R, 4 X$ watts into $2 R$ and $\infty$ watts into a short circuit! All without blowing fuses of course.

Allied to the good and true RMS readings, for long term delivery, I think these burst measurements paint a fairly good picture of the capabilities of a particular unit with regard to actually playing records.

I would welcome readers' comments on the method outlined so far, or any suggestions as to how it could be improved. (Physically possible suggestions only please, it's a big test-bench.)

## Hitachi Hi-Power

In the bridged mode the 7500 II delivered 150 W into 8 R all the way from $20 \mathrm{~Hz}-20 \mathrm{kHz}$. Burst power was a very healthy 201 W into 8 R and 189 W into 4 R . It is somewhat more choosy about impedance in this mode, so I did not attempt a 2 R reading. Test loads are expensive and I'm running out of fuses.

Signal-to-noise was an impressive 94 dB down, unweighted and damping factor held above 50 down to 40 Hz . Provided the speaker doesn't take a downward leap in impedance, therefore, good control of the bass could be expected.


Bridging the 7500 II means connecting across the 'lives' of the two sets of speaker terminals. This is denoted by the obscure legend BTL. It actually stands for "Bridge Transformer Less"! I kid you not.

## Testing Conclusions

Overall then, the combination gave a good technical performance, with the power amp appearing to be of a higher standard than the preamp, with only a very few reservations.

In order to test both, the combination was first tried together and then independent of each other, with known preamp and
power units standing in for the missing partner where necessary. In this way a clearer idea of the sound behaviour of each unit could be obtained.

## TEST RESULTS

HCA - 7500 II


## Making MOSFET Music

As a pair the 7500 s performed well - with the previously stated reservation on pickup input - sounding clear and powerful. Mid-range rendition was good and bass transients well handled. The treble tended to be a little 'soft', if the unit is driven hard - an effect which is infinitely preferable to the biting edge some designs reveal under these conditions.

The preamp has a good performance for the price, which is low enough that, should you upgrade your cartridge beyond its capabilities, you could afford a specialist pickup amplifier to overcome the differences.

The power amp is very good indeed. It separates out the components of a complex piece very well, and portrays a convincing stereo image. The wildly dancing meters are impossible to ignore - what that means to you, be it good or bad, is a personal decision.

In bridged mode - which I have only been able to try in mono as yet - I felt the amp was even better, more open and with a more confident bass. I have contacted Hitachi and hope to be able to include a bridged pair of 7500 s in the forthcoming comparison.

## Final Notes

Final comments? Very good power amp by any standards. The preamp is good value for money, but the pickup input lets it down slightly. Overall a worthy production and one that cannot be considered expensive at $£ 400$.


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# ROBOT MOTOR CONTROL PART 3 

# This month we describe the construction of the analogue pulse width modulator and show how to interface it to the dual motor controller. Design and development by Rory Holmes. 



The overlay diagram for the analogue pulse width modulator described last month is shown in Fig. 2. Assemble the PCB following the component orientations indicated and soldering in sockets for the ICs. If only one channel is required, the board can be cut in half along the dividing line, omitting the two links and assembling the circuit for channel 1. Overlay pictures for the transistors are correct for the specified ' $L$ ' versions and pin outs should be identified carefully if other types are used.

Veropins should be soldered in at all the points marked for terminations (18 in all). Spare positive and ground terminals have been included near the input and output points for flexibility. The resistor shown on the circuit diagram in last month's issue as the speed feedback input and marked as Rx on the overlay, should be included on the board; a value of 100 k is
required. A 1 M 0 shorting resistor should also be connected between the terminal pins marked for speed feedback input (GND and INPUT). This resistor prevents stray interference in the basic unit and can be easily removed for later addition of closed loop velocity control.

After mounting the two presets they should be set at about half-travel. (Ceramic base presets can crack quite easily, so take care when inserting these into the PCB.) These presets are used to apply a DC offset to the voltage summing amplifier, so shifting the control voltage range to the required level.

## Installation

Without plugging in the ICs , a 12 V power source can be connected to the board as shown. The power rails should be checked at the relevant pins on the IC sockets; also check that the 8 V
reference is present at the positive end of C13 and C6. If all is well, the power can be disconnected and the ICs plugged in. At this stage the board can now be mounted using brackets or insulating pillars directly above the power switching PCB in the diecast box. The 12 V power source is taken from the existing bridge-rectifier tags, and the two output signals, PWM and FOR/REV, should be wired to the corresponding pins on each channel of the power switching PCB. Have a look at the internal photographs with the analogue board.

The two manual control potentiometers, RV1 and 2 can now be wired up to the control voltage input. In the circuit diagram the positive ends of RV1 and 2 are shown connected to the +8 V reference; a four core cable would thus be needed for the remote attachment of these potentiometers with their limiting resistors R8 and R22.


Fig. 1 Wiring diagram for the manual control unit.


The analogue PWM board installed above the dual motor driver.

Essentially all that's required is a voltage variable from 0 to 8 V applied to the base of Q3, so to allow the use of cheaper twin core screened cable we mounted our pots in a small handsize box with their own 9 V PP3 battery power source. A five pin DIN plug and socket mounted at one end connects the control voltages from the pot sliders, via the cable, to the modulator PCB. The cores are wired to the input terminals at the bases of Q3 and Q6, with the screen taken to the adjacent ground pin. Figure 1 illustrates the manual control wiring, the accompanying photo the internal appearance.

R8 and R22 are shown on the circuit diagram as 1 kO ; these may be altered as required to limit the maximum reverse speed. RA and RB however are optional resistors for limiting the maximum slider voltage to 8 V ; we used 1 k 0 resistors to suit the 9 V battery.

If this type of manual forward/ reverse control is not required the associated components can be omitted during assembly (C4, C11, Q3, Q6, IC2, IC5, R9, R23, R10, R24). Pin 6 of sockets IC2 and 5 now becomes the FWD/REV logic control, the unused collector pad of Q3 and 6 becomes the control voltage input.

The ratio of R15 to R13 determines the gain of this input, and with the specified 1 M 0 values is set at unity. Thus an input variation over a 3 V range will provide $0 \%$ to $100 \%$ dutycycle control. The input should be referenced to the 0 V ground, and PR1 and 2 can provide full offset adjustment of the control ranges. To increase the gain, the value of R13 should be decreased.

## Testing

Once the manual potentiometers have been wired up in a suitable fashion the completed controller may be set up for proper operation. Temporarily disconnect the PWM signal wires and solder them to the adjacent ground terminals; this will prevent the power stages being damaged if there are any errors.

Connect a 12 V supply to the main controller; two glowing LEDs are the first signs of success! A voltmeter or scope set for 12 V FSD should now be hooked up to the FWD/REV output; depending on the pot position a 0 V or 12 V level will be present, and should sharply change state as the pot is turned through its centre travel. A ramp waveform of 3 V peak should be observed if a scope is put on pin 6 of IC3 or IC6. The PWM output can now be measured for each channel with the meter, indicating a voltage


The manual control unit.


Fig. 2 Component overlay for the analogue PWM unit.
proportional to duty-cycle, or a scope to show the pulse waveform. For a given position of the control pot, clockwise adjustment of the corresponding preset will increase the duty cycle. Leaving the preset and now turning the manual pot towards centre travel will decrease the duty-cycle linearly until the pulse width vanishes, giving a dead band of 0 V . Turning the presets again, anticlockwise this time, will increase the deadband. A small proportion of the pot rotation should be left as deadband, so enabling the motor stop position to be easily located. Adjustment is more difficult with a meter than a scope, but with patience the desired setting can soon be achieved.

The PWM signal wires can now be reconnected to the outputs and the chosen motors wired across the power amplifier load terminals. A 20 A toggle switch wired in series with the positive battery supply is strongly
recommended at this stage. (WARNING: If the motors used are rated at less than 6 V , the maximum obtainable duty-cycle should be limited accordingly by increasing the deadband; good motors are usually expensive.) A 12 V bulb will provide a good motor substitute for the purposes of testing. Each manual pot will now independently control the average power into any load; bidirectionally!

## PARTS LIST

## MANUAL CONTROL BOX

Resistors RA and RB; $\mathbf{1 k 0}$ ( $1 / 4 \mathrm{~W}, 5 \%$ ) (see text) Case (see Buylines)
5 pin DIN plug and socket
PP3 9 V battery
Twin core screened cable (length to suit remote operation)
Two control knobs

## BUYLINES

You shouldn't have any supply problems with the components for this project - everything is absolutely standard. The case we used is a Vero type $65-2514 \mathrm{~F}$ and should be available from any Vero stockist. We can supply the PCB - the order form is on page 45.

Coming next month: not only the digital PWM board for direct computer control of the motors, but an infra-red proximity detector too. We also begin to assemble the modules into a working mobile.

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# ENGINEER'S GUIDE TO PRINTERS 

# Whether you type LPRINT or OPEN 1, 4, not much is going to happen unless you've got a printer hanging off the back of the computer. Robert Traub helps you decide which one is right for you. 



Printers vary greatly in their functions and features. There are a great number of types to choose from and the choice can be almost impossible. Some very basic points about printers may assist in the selection of a printer for personal use. The first point to cover is columns. What is a column? Some printers will print 132 columns, some only 40 columns and some just about everything in between and more. A column is the space occupied by a single character. Consider first printers with fixed pitch. Fixed pitch means that each character occupies the same amount of space on the page. The common. fixed pitch is 10 characters per inch; this standard pitch would allow 85 columns across an $81 / 2$ inch wide paper if no room is allowed for margins.

If the page were allowed to have margins of approximately $5 / 8$ inch on each side, that would leave 7.2 inches. At 10 characters per inch, that would give us 72 columns, and this is the standard print page for TTY type printers (and others). If the printer were to offer 132 columns, then the paper would have to be at least 10 characters per inch divided into 132 columns equals 13.2 inches wide. If margins were to be included it would bring the width of the paper to $151 / 8$ inch. Therefore a common 132 column page would be $15 \%$ inches wide by 11 inches deep.

Some fixed pitch printers offer the ability to select the pitch at which the characters will be fixed. Some common values are 12 characters per inch and 13.5 characters per inch. A bit of math would soon tell us that a printer with a fixed pitch of 13.5 characters per inch could print up to 96 characters or columns on a standard page with margins, while a printer with a fixed pitch of 12 characters per inch could print 87 columns in the same space. The 13.5 character per inch printers compress the

characters much closer together and may be a bit harder to read if they don't use a good quality print head. Some dot matrix printers offer a very compressed print of 132 columns in eight inches or 16.5 characters to the inch.

Next we will look at printers that are not fixed pitch. A few printers offer the ability to allow a given amount of space on the line to each character. With this type of printer, less space would be allowed for the letter " i " than would be allowed for letters such as " $m$ " or " $w$ ". Each letter is assigned a given amount of space on the line, as well as the distance between each letter. (With fixed pitch printers the amount of space allotted is the same for all characters.) This type of print is referred to as proportional spaced print.

With this type of print you can develop excellent quality documents and avoid the common "river of white" that is found running through the body of text done with fixed pitch printers. Proportional spaced printers are the type required for professional word processing systems; the quality of print is excellent, and the overall appearance of the documents produced on the system is outstanding. These types of printers are very expensive, and very elaborate software is required for their operation in order to exploit their full potential.

Before leaving this subject, there are a couple of terms that you may run across and not be sure of their meaning, these being pica and elite. The term pica refers to typewriter type providing 10 characters to the inch, while elite is typewriter type that provides 12 characters to the inch.

## Type Of Type

The next thing to consider is the type of print. One common type is dot matrix print. This type of printer comes in many dot matrix forms; some may be 5 by 7 , some 7 by 9 , some 9 by 9 , and some even greater. The dot matrix is comprised of a number of small wires or pins that are struck against a ribbon and the paper leaving a dot on the paper. The number of dots that the matrix has will determine the fullness of the character that it is reproducing. A $5 \times 7$ dot matrix is comprised of 35 small dot positions arranged in such a manner as to have a total of five dots across (horizontal) and seven vertical rows of five dots each. This arrangement is the minimum number of dots required to produce a decent upper case letter set and numerals. The characters are not always well formed as can be seen by the " S " not being curled around at the top and bottom, and by such characters as the slash(). The slash will have a small vertical line on both ends rather than being a single straight line.

A $7 \times 9$ dot matrix is arranged in such a manner as to have seven dots across (horizontal) and nine vertical rows of seven dots each for a total of 63 dots. This type of matrix will produce much better letters, and give a more natural look to the overall print. As the number of dots in the matrix increases, the ability to reproduce characters increases and some very nice naturallooking print can be found with such printers. Of course, if graphics are being considered, the greater the matrix dot count, the better the quality of graphic representation. Better quality print will be produced by matrixes with greater numbers.

The least expensive of the dot matrix printers will generally have a standard 5 by 7 matrix print head. This printer is satisfactory for general use, but is not intended for word processing as it does not have descenders. A descender is the tail of lower case characters such as ' $p^{\prime}$, ' $q$ ', ' $j$ ', ' $y$ ', etc. Note that the tail of these characters will extend below the base line on a normal typewriter quality printer. On dot matrix printers this is not always available, and never on a $5 \times 7$ dot matrix. As the number of dots in the matrix increases, so does the price and overall general quality of the printer. The very elaborate dot matrix printers that are now available can rival almost any type of print, but are very expensive and therefore not generally appropriate for hobby applications.

Printers that offer fully-formed characters such as found on typewriters are the best quality for word processing at a more reasonable cost. Fully formed characters are the type found in typewriters and are cast in metal or plastic. Some systems have the characters formed on a ball, some have them formed on a cylinder and others use a daisy wheel, to name but a few methods.



The daisy wheel comes in different sizes with 88,92 or 96 characters per wheel and can be cast in either plastic or metal. Some of the cheaper printers, whether dot matrix or full formed character type, do not offer lower case characters; again, this may or may not be important to the user, although lower case is a must if word processing is being considered. Each printer must be studied in order to determine if it offers lower case characters, descenders and other features (graphics) that would be of interest to the user.

## Feed For Thought

This brings us to the question of friction feed or tractor feed. In the case of friction feed, the paper is held in place by some small pinch rollers that press against the printer's platen. The paper is inserted between these pinch rollers and the platen. This is fine in most cases where each line is advanced one at a time by a carriage return-line feed combination, but if the lines were to be advanced by the inch with a sudden command, as is the case with the form feed character, the paper could 'slip' as the platen first starts its fast advance. To overcome this problem, the tractor feed type of paper advance systems can be used. With this type of printer option, paper can be advanced rapidly with assurance that the print will start at the same line on each page or form.

One other type of paper feed system is the pin feed; this system is used on TTY printers to ensure that forms such as telegraphs will always line up properly. Another feat Ire offered by the tractor feed and pin feed systems is the assurance that the printed line is horizontal with respect to the top and bottom edge of the paper. With friction feed systems, the page can slip slightly one way or the other.

As there are different systems that can be used to feed paper, the choice will depend on the type of work the printer will be required to do. If a lot of forms are going to be filled out, then of course a printer with the tractor feed option would be a good choice. If individual letters are the order of the day, then the standard friction feed type of paper advance will serve well.


## BAUDy Stories

Briefly we will take a look at the question of baud rates. The baud rate or just plain baud means number of code elements per second. If the ASCII code were taken as a 10 unit code, then a baud rate of 1200 baud would transfer data at the rate of 1200 elements per second divided by 10 units per character for 120 characters per second. If the ASCII code were to be considered as an eight unit code, then 1200 baud would represent a rate of 1200 elements per second divided by eight units per character for total of 150 characters per second. Many printers will accept data at a rather high baud rate, say 9600 baud; this is the rate at which they accept data into their buffers and not the rate at which they print. The printer may only be able to produce 150 characters per second on paper; therefore, that printer's true baud rate is 1200 baud if an eight unit ASCII code is assumed. The baud rate then is the speed at which the data or information can be printed.

There are many reasons why a faster speed is needed in some cases and not at all needed in other cases. Typical baud rates range from a slow 110 baud to a fast 9600 baud, but be sure to check if the baud rate is the rate that the printer will print characters or if it is the rate at which the host computer can send characters to its internal storage area (buffer).

## Drop Me A Line

Some printers you hear about are called line printers; a line printer is a special type of printer that will not print each character as it is received, but rather will wait for the complete line and then print the entire line at once, a character at a time. The length of the line that it prints is determined by the sending of the RETURN character, as a return signifies the end of that line in text. Line printers require special handling by the host computer and provisions must be made for handshaking. Line printers have buffers to store the data in before it is printed and handshaking is simply the printer's method of telling the computer when to send more data to the buffer and when to stop sending data as it cannot handle any more it the moment. Printers are generally slower than the host camputer, although there are some very fast printers not generally used by the hobbyist.

One other thing that you might run across is the term $b \dot{F}$ directional printer. What this means is that the print head will print a line from left to right across the paper, advance the line (line feed) and then print the next line from the right side back to the left. The bi-directional printer requires fewer mechanical parts and movements than single direction types and this is one reason for increased printer speeds. With the conventional type of printers, a carriage return is required in order to bring the print head back to the start of the next line. This takes time and computers have to send the printer a pad or fill character in order to assure that the head has returned to the far left before it starts printing again. After many, many line feeds and carriage returns the amount of time wasted can be considerable. Therefore the bi-directional printer, which does not return the head every line, is capable of greater speeds (throughout). Because there are fewer mechanical parts to wear out, reliability is also increased


## Summary

At this point there are a few things to consider. If the printer is going to be used for listing BASIC or assembly programs, the dot matrix type of printer or the teletype printer may do the job very well. If a teletype printer is used, you will have upper case only, a slower 110 baud rate and require a 20 mA loop interface. Printers of the dot matrix type vary by a great amount and will require either a parallel or RS232 serial interface.

The lower priced $5 \times 7$ dot matrix printers have a great deal to offer those wishing listings. They can be found in a variety of speeds ranging from 110 baud to 9600 baud, and vary greatly as to the size of paper that they can accommodate. If you require listings at 10 characters per inch on an $11 \times 8.5$ sheet of paper ( 11 inches horizontal and 8.5 inches vertical), you may find that the printer will not accommodate this size of paper and, in fact, the largest that it can handle is 9.5 inches horizontal.

If the printer cannot handle wide paper, check out the print quality of the printer at its most compressed setting, 13.5 characters per inch for instance, and use a very complex BASIC program line with as many characters and commands as possible. This will allow you to see first hand the type of printout you will be trying to work with. You may find it unsatisfactory for long listings, as the heavily compressed text makes it very hard to find semi-colons and other required BASIC syntax characters. If you will be doing a lot of amortization charts, or charts of any type, then it is recommended that you look at a printer that falls into the line printer category, as the speed will be needed. You would soon go broke trying to produce amortization charts at 110 or even 300 baud.

There are as many reasons for getting a printer as there are printers on the market; it is advised that the first thing you do when thinking about getting a printer is sit down in your computer room and try to determine exactly what service you want that printer for, and then proceed to find one that will do those limited number of things. You can never be sure that you have covered all the bases, so be sure to explain to the dealer what use you wish to put the printer to, and he will be in a position to advise you as to what you might need that you have overlooked. Try to stay away from a printer that will do everything; they require special features and will do all the different things if you buy a lot of extras, for example.

Some printers offer tractor feed as an option; this may be a valuable option to have, as trying to stop the printer at the right time to change paper for the next page is not always desirable. Check your computer program and see if you can stop the printer from the terminal keyboard at any given time, or if the software will do that automatically. If not, you will need tractor feed and continuous form paper for the job you are doing or roll paper if friction feed. If the printer is to be used for word processing, even on a small scale, it is recommended that you have lower case with descenders at the very least. Speed is not important in this type of printer and print rates of 30 to 55 characters per second are common. What is important is the quality of print and that is what you should look at first and last.

There are other things to consider in purchase of a printer, such as whether a warranty or service contract is available. Second-hand equipment often does not come with service contracts. The application the printer will be used for will largely determine the quality of print required, but the cost could be the main consideration for the hobbyist.


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## ELECTROMUSIC TECHNIQUES

## The second in this design series by leading-expert-in-the-field Tim Orr features voltage controlled filters, voltage controlled amplifiers and ring modulators.

The first group of circuits we corisider this month are voltage-controlled filters. Figure 1 shows the circuit for a state variable filter with four frequency responses; lowpass, highpass, bandpass and notch. All four responses can be controlled by varying the gain of the two integrators. The Q factor of the filter can also be voltage controlled. If the $Q$ is set to maximum, by turning off the feedback CA3080, then the circuit will become a sine wave oscillator (because the damping has been reduced to zero). Prior to this, very high Q factors can be obtained, of the order of 400 . The frequency responses are shown in Fig. 2. Most synthesisers use a $-24 \mathrm{~dB} /$ octave lowpass VCF, but the more responses that are available, the wider is the choice of sound that can be produced.

VCFs are usually swept with a control voltage from an ADSR. Every time a note is played on the keyboard the VCF is swept, the shape of the ADSR signal and its polarity determining the type of sound that is heard. Figure 3 is a circuit for sweeping a VCF. Both positive and negative sweeps are obtained on one control pot. Curtis make a VCF chip, the CEM3320 (Fig. 4). The configuration shown is a four pole ( $-24 \mathrm{~dB} /$ octave) lowpass filter with $a+1$ V/octave sensitivity and voltage control of the Q factor.


Fig. 2 State variable filter responses.
Fig. 1 State variable VCF.

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Fig. 3 (Below) VCF sweep control unit.



Fig. 6 (Above) Biquadratic VCF.



Fig. 7 Biquadratic lowpass response.

Another CA3080 VCF is shown in Fig. 5. Note that the accuracy of the exponentiator need not be as good as that needed for a VCO (unless you are going to make the VCF oscillate and track a VCO). Therefore the Q81 resistor has been left out. A somewhat different VCF is shown in Figs. 6 and 7. The biquadratic filter has a Q factor that is proportional to the cut-off frequency. So, as this frequency is increased, the $Q$ factor will increase. This gives a constant ringing time which is independent of frequency. All the previous VCFs have a constant $Q$ operation.

## VCA And Ring Modulators

Voltage controlled amplifiers are one of the easier synthesiser building blocks to make, as long as you don't want low noise and low distortion operation. Figure 8 shows a standard linear VCA. The audio input is attenuated to about 40 mVpp and then fed into the CA3080, the gain of which is controlled by the $I_{A B C}$ current. If the audio input is removed, control breakthrough will probably be seen at the output. Most of this is caused by the input offset voltage of the CA3080 being multiplied by the $I_{A B C}$ control current. The offset can be nulled
out by adding a small DC voltage to the non-inverting terminal, which should eliminate most of the control breakthrough Any residual breakthrough is due to current mirror mismatches in the CA3080 and is unavoidable. Distortion may also be rather high, perhaps in the region of $0.5 \%$, but this is not generally considered to be a problem in synthesiser circuits. Lowering the input signal level will reduce the distortion at the expense of an increase in the noise level.

A better VCA is shown in Fig. 9, the CEM3330 made by Curtis. This is a dual device with both $\log$ and linear control inputs, low noise and low distortion plus very low control feedthrough. A third VCA is shown in Fig. 10, this one being constructed from a CA3046 transistor array. It uses two of the transistors as a predistortion circuit so that a higher operating signal level can

be used for the same level of distortion. In fact, the predistortion principle is used in several multiplier chips, including the LM13600 which is used in the next circuit, (Fig. 11). The two LM13600 circuits are used a low distortion VCAs. A predistortion diode bias current is inserted into the IC at pins 2 and 15 . The gain of each VCA is controlled by the $I_{A B C}$ current (pins 1 and 16), this current being derived from a pair of complementary control voltages. As the gain of the channel increases the other decreases. Some interesting effects can be obtained with this circuit; for example a note can pan from left to right every time it is played.

The VCAs mentioned so far have all been two quadrant multipliers. The operation of a four quadrant multiplier (sometimes called a balanced modulator or ring modulator) is very different (Fig. 12). When two sine waves are multiplied together the result is a signal composed of sum and difference tones. For example, if the two input sine waves have frequencies of 100 Hz and 1 kHz , then the output will be composed of two tones, one at 1100 Hz (sum) and one at 900 Hz (difference). If the same sine wave is applied to both inputs, then the sum tone is twice the original frequency, and the difference tone is a DC voltage. Ring modulators are used to produce discordant sounds and special effects such as the BBC Dalek voice.

Figure 13 is a simple ring modulator circuit. The performance suffers a bit from poor $X$ and $Y$ feedthrough, which can be minimized by adjustment of the two presets. A better modulator is shown in Fig. 14; this circuit employs a balanced modulator chip made by National Semiconductor and others.

Fig. 9 (Left) Monolithic VCA circuit.
Fig. 8 (Above) Simple VCA.


Fig. 10 (Below) Discrete VCA.


The feedthrough adjustments are very sensitive and so it is necessary to run the circuit from a stable pair of supply rails. Adjustment of the presets is as follows. Insert a carrier signal ( 1 kHz at 2 Vpp ), look at the output and adjust the carrier fundamental and then the carrier second harmonic presets for minimum feedthrough. Repeat this for the signal path. Feedthrough should be the order of 60 dB down on the maximum

Fig. 11 (Above) Voltage controlled panning.


Fig. 12 (Above) Ring modulation.


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Compared with conventional instruments this oscilloscope is smaller, lighter and better value for money. It is barely larger than a house brick and about as heavy as a good multimeter; thus it is ideally suited to mobile usage on building sites, in installation work, in motor vehicles and so on. The instrument can be powered either from the mains, using a separate 240 V to 12 V transformer, or direct from a 12 V battery. The specifications are given in the table.

The individual circuit blocks are connected by ribbon cable. The 12 V DC or $A C$ supply enters via two connectors on the rear of the instrument, passing through a two-pole switch (linked to the X-position potentiometer) to the power supply card. After stabilisation to 10 V , the various working voltages are generated here using a 25 kHz switching circuit. Stabilisation of the input has the advantage that all the output voltages are thereby simultaneously stabilised, ie they are independent of supply variations. Because a switching regulator represents a voltage source having a low internal resistance and the current output to the individual circuit blocks is largely constant, further secondary stabilisation measures are unnecessary.

Also on the power supply card are the individual trimmers (focus, astigmatism, brightness) and the voltage multiplier for the tube supply, so that the tube (up to the deflection plates) can be connected directly from here by means of a six-way ribbon cable. The connections for the tube are made direct to the base. No socket is used, because the additional capacitance would considerably reduce the bandwidth.

A seven-way ribbon cable connects the main circuit board with the various supply voltages, and the


Now this is what we call small.....
(flyback) blanking pulse goes from the main board to the grid G1 of the tube. The main board contains the trigger switching, the sawtooth generator and the X - and Y -deflection amplifiers.

The three potentiometers mounted on the front panel ( X - and Y -shift, trigger level) are connected by a five-way ribbon cable to the main or preamplifier boards, as appropriate. Similarly the time-base selector on the
front panel is connected by three single wires to the main board.

The input voltage divider forms a complete screened assembly, comprising switch SW1 and the voltage divider and preamplifier boards - thus no additional external wiring is required. The complementary output of the Y -amplifier passes to the $Y$-deflection amplifier on the main board.

## SPECIFICATION

Bandwidth: 0-7.5 $\mathrm{MHz}(-3 \mathrm{~dB})$ for six divisions (one $\operatorname{div}=7 \mathrm{~mm}$ ) $0-10 \mathrm{MHz}(-3 \mathrm{~dB})$ for four divisions.
Input: BNC connector, switchable $\mathrm{AC} / \mathrm{DC} /$ ground.
Sensitivity: $5 \mathrm{mV} /$ div to $20 \mathrm{~V} /$ div in 12 calibrated $1 / 2 / 5$ steps.
Case Size: approximately $175 \times 105 \times 100 \mathrm{~mm}$.
Weight: approximately 1 kg .




The tube is supported by its metal screen and cushioned by foam draught excluder. The power supply is mounted beneath it.

## HOW IT WORKS

## VOLTAGE DIVIDER <br> AND PREAMPLIFIER

The input signal reaches the input voltage divider from the BNC connector and the mode selector SW2. After passing through the voltage dividers, the signal goes via R3 (over-voltage protection for Q1) to the gate of the dual FET Q1. The second gate of this transistor is connected to the Y-shift potentiometer RV1 and permits vertical shifting of the 0 V line on the screen. Any mismatch between the two halves of Q1 can be trimmed out by means of trimmer PR1. Q1, which is connected in the source-follower mode, serves merely to buffer the high impedance of the instrument input and the voltage divider from the low input impedance of the preamplifier IC1. The gain is fixed and determined by R12. Complementary output signals appear at $Q$ and $Q^{\prime}$ and pass along single wires from here to the $Y$-deflection amplifier on the main board. MAIN BOARD
The two transistors Q2 and Q3 on the left of the diagram take care of the final amplification of the $Q$ and $Q^{\prime}$ signals provided by the preamplifier. The amp; lified signal is connected to the $Y$ and $Y^{\prime}$ deflection plates using short, single wires (lowest possible capacitance!) going directly to the base pins. The working point of the stage is adjusted by PR3, the gain by PR2. C10 and CV10 serve to linearise the frequency response.

The trigger signal is taken from Q1 via C8. Q4 and Q5 form a preamplifier whose working point is adjusted by the trigger level potentiometer, RV2. Q5 feeds the Schmitt trigger IC2a which con-
verts the input signal into rectangular pulses. These rectangular pulses are differentiated into 'spikes' by C17 and R43, and are fed to the 'set' input of the gating flip-flop (pin 4 of IC26). The gating flip-flop enables the sawtooth generator for one beam-sweep (in a horizontal direction on the tube).

In order that a 0 V line will be traced in the absence of an input signal (and therefore in the absence of the 'spikes') - for example, for direct voltage measurement, or for reference purposes with the input switch on ' $G$ ' - transistor Q6 is switched into conduction and thus gives a continuous 'set' condition. As long as rectangular pulses are present at the output pin 3 of the Schmitt trigger, they will be rectified by D1 and D2, generating a negative gate voltage for Q6, so that this stays non-conducting. Thus the 'Free-run/Automatic' Q6 only works when there are no trigger pulses available.

The output ( pin 8 ) of the gating flipflop switches on Q7. This provides base current - through R50 and PR4 - to transistor Q8, which is connected as a constant current source. The magnitude of the constant current which Q8 provides is determined by emitter resistors R44 to R49 (selected by the time base switch SW3). As soon as Q8 is switched on (by Q7) one of the two capacitors C19 or C20 will be charged (according to the position of the time base selector). Since this charging takes place from a constant current source, the voltage on the capacitor increases linearly with respect to time. This linear (sawtooth) voltage in-
crease is fed to the X-deflection transistors Q12/Q13 by Q9 and Q10, so that the X and ' X' deflection plates receive a linear beam sweep (from left to right on the screen). A part of this voltage is fed via R54/R55 to the input pin 12 of the reset trigger (IC2d). As soon as the sawtooth voltage attains such a level that the beam reaches the right hand side of the screen, pin 11 of IC2b switches low (approximately 0 V ). This switches Q11 on, and also supplies G1 of the tube (via C24) with a negative-going, 40 V pulse, which blanks the beam. At the same time, the timing capacitor (C19 or C20) is discharged, which returns the beam (blanked during the flyback) to the left-hand side of the screen. Pin 11 of the reset trigger also drives the 'reset' input (pin 10) of the gating flip-flop. This switches off Q7 and Q8. As a result there will be no new beam sweep for the moment, while the current source Q8, which charges the timing capacitors, is switched off. A new sweep will only be initiated by a trigger pulse on the 'set' input pin 4, and this will occur at exaotly the same point on the waveform of the input signal as for the preceding sweep. This ensures that successive traces in a continuous sequence are written on the screen in a uniform manner, provided that the input signa! is not altered. Trimmer PR4 provides for calibration of the timing circuit (charging current adjustment), PR6 is the working point adjustment, and PRS is the gain (picture width) control for the X-deflection amplifiers. The trace can be shifted from right to left on the screen, from the front panel, by RV3.



Fig. 1 Circuit diagram of the input circuitry. The voltage divider is built around switch SW1 and includes the four divider blocks which are drawn separately.

The component overlays and constructional details for the oscilloscope will be given next month.


Fig. 2 Circuit diagram of the main board. The two dotted capacitors in parallel with C19 and C20 may be required for trimming purposes - this is covered next month.

Fig. 3 Circuit diagram of the power supply.

> NOTE: Q15 IS BC252C O1618,19 ARE BD135 Q17'IS BC 172C DA.7 ARE 1N4002 D8.17.21 ARE 1N4 148 D9. 12 ARE BA158 D13.16 ARE BAV20 ZD1 IS 5V6 ZENER ZD2 IS 47V ZENER


Fig. 4 Block diagram of the oscilloscope.


## HOW IT WORKS

## THE POWER SUPPLY

The incoming supply voltage from SW4 (coupled to the X-position potentiometer) and fuse FS1 is rectified by D4-7. These also ensure correct polarity of the input voltage in the case of a DC source. The supply voltage is stabilised to 10 V by Q15 Q16 and Q17. The constant and almost ripple-free voltage then passes to the transformer T1 and the 25 kHz oscillator formed by Q1 8 and Q19.

R73 limits the base current to Q18 and Q19. R74/C28 suppress switching voltage spikes. R75 in conjunction with D8 acts as a starting circuit (for the oscillation).

The tube heater voltage is generated by the transformer secondary winding T1c. Winding T1d delivers a 156 V square wave which is rectified by diodes D13-16 and passed to the deflection amplifiers. Diodes D13-16 must be BAV20 and D9-12 must be BA158, as specified in the Parts List; on no account can ordinary bridge rectifiers or rectifier diodes (eg 1N400X) be used, as these are an order of magnitude too slow to rectify a 25 kHz signal.

The 156 V is reduced to 100 V (for the Y-deflection amplifier) by R78/R79. From this is derived, through PR10, the tube G2 voltage (astigmatism adjustment) and also the 40 V DC for the flyback switching, via R82 and ZD2. All the voltage values given in the circuit diagram are measured with respect to ground. T1d also feeds the voltage multiplier (D9-12, C29-32) for the generation of the tube EHT supply of -460 V . This voltage is fed to a potential divider (PR8 - brightness, PR9 - focus, R80) on which the individual electrode potentials for the tube are available.
The +7 V and -6 V supplies are generated by T1e. The asymmetry of these voltages is a result of the different values of resistors R76 and R77. This is necessary for the supply to the preamplifier IC1, whose inputs, pins 1 and 14 , sit at about +1 V. The supply (source voltage of Q1) is therefore sufficiently compensated that the positive voltage (on pin 10) is somewhat greater than the negative voltage on pin 5.


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## TECH TIPS



Fig. 1 Circuit of one channel of the CMOS vocoder.
Fig. 2 Several channels are connected in parallel to form the complete vocoder.


## Economy CMOS Vocoder

S.P. Giles, Edmonton

The principle of the vocoder has been well covered in ETI in the past but it is very expensive to construct such a project due to the high component count, especially op-amps. This circuit attempts
to bring the vocoder nearer the pockets of those who would not normally have sufficient funds. CMOS inverters are used instead of op-amps for cheapness.

The signal to be analysed, normally the human voice, is fed into the mike input and amplified by inverter IC1a. IC1b, a bandpass filter, allows a very narrow band of the input through. The centre frequency of this band is determined by PR1. The amount of energy present in this narrow band is measured by an envelope follower consisting of inverters IC1c, d and D1.

If a sound source is connected to the other input it is necessary to transfer onto it the energy present at point A . This is done by Q1, a voltage-controlled amplifier with the level set by the envelope follower voltage. PR3 should be adjusted for zero output at point $B$ with no input. Inverter IC2a is a simple mixer for all the VCA outputs.

The second diagram shows how the filter bank is built up. It is suggested that a minimum of 20 channels be used to get the best results. In this case PR1 and PR2 should be set to approximately 1 k 2 intervals from one channel to the next. The greater the number of channels the better the results.



## Guitar Tracking Oscillator

S. P. Giles, Edmonton

This circuit will enable a guitarist to produce a second note in harmony with the note he is playing. IC1a amplifies and squares up the incoming guitar signal which is then passed through 1 kHz lowpass filter IC1b. This is to remove most of the harmonics leaving a reasonable square wave at the fundamental pitch of the note played. IC2a, 2 b and IC3 form a simple compressor so that phaselocked loop has a better chance of locking on to the square wave applied to pin 14. Inside the 4046 , there is a VCO which oscillates at a frequency
determined by the RC network around pin 9 , and is applied to one input of a phase comparator. An external square wave applied to the second input will produce an error voltage which is proportional in magnitude to the difference between the external and VCO frequencies. The error voltage is passed to the VCO until the two frequencies are the same, ie the VCO is locked to the input signal.

The error voltage can be extracted from pin 10 of IC4 and passed via RV2 to a second 4046, IC5. RV2 varies the level of the DC signal which controls the frequency of the VCO in IC5. The resistors from pins 11 and 12 of IC5 can be changed to alter the maximum and minimum frequencies of the internal VCO.

## Simple Graphics On A Scope Screen

G. Heath, Oxted

The circuit described here will display eight 16 -bit words on an oscilloscope screen. The counters and D-to-A converter provide positioning information which is scaled by the two 741 amplifiers. The actual 1 s and 0 s are produced by the summing of a sine and cosine wave to the X and Y amplifiers. When the 74150 outputs a 1 the $Y$ output is shorted to ground, drawing 1 on the screen.

The size of the characters can be ad-


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| Sx18 | 100 | Mixed Ceramics Lpf. 5 Gipf | 1 |
| $5 \times 19$ | 100 | Mixed Ceramics 6apl 0.5 mf | I |
| SX20 | 100 | Assorted Polyester/Polystyrene Capacitors | + |
| SX21 | 50 | Mixed C280 type capacitors metal toil |  |
| 5x22 | 100 | Electrolytics, all sorts | C1 |
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# SOUNDEFFECTS 3 

## The SN76488 is pretty versatile. So is our sound effects PCB. Hence their return this month with two new noises to astound and amaze, the first being a phasor and explosion. Design by Phil Wait.

This combines a 'phasor' effect and the explosion effect employed in the Bomb Drop and Explosion unit last month. One could liken the sound produced to what you would expect after shooting down a 'flying saucer' or somesuch! This project uses about as many components as the Bomb Drop and Explosion board.

The SLF sweeps the VCO up and down in pitch at quite a rapid rate the push-button is held down to start the effect, which takes several seconds to complete. The explosion is heard following a period of the phasor sound As with the other units, if you wish to vary any of the parameters of the effect it is best to vary the resistor values.

Take care with the orientation of the electrolytic and tantalum capacitors during construction. Note that, as with the Bomb Drop and Explosion unit, there are two links on the board; make sure you don't miss the small link at the 'notch' end of the IC.

PARTS LIST

| Resistors (all $1 / 4 \mathrm{~W}, 5 \%$ ) |  |
| :---: | :---: |
| R1,5 | 470k |
| R2 | 100k |
| R3 | 150k |
| R4 | 220k |
| Capacitors |  |
| C1,2,5 | 4 u 716 V PCB electrolytic |
| C3 | 4 n 7 polycarbonate |
| C4 | 470p ceramic |
| C6 | 100u 16 V PCB electrolytic |

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Miscellaneous
PB1 SPST push-button switch
PCB (see Buylines); $\mathbf{5 0} \mathbf{m m}$ diameter $\mathbf{8} \mathbf{~ o h m}$ speaker; PP3 battery and clip.

## BUYLINES

The SN76488 is available from Technomatic at a price of $£ 3.50$, while the PCB (identical to last month's one) will cost $£ 1.80$ from our PCB Service (see page 45) if you don't want to etch it yourself. Everything else is pretty run-of-themill.

Fig. 1 Circuit for the Phasor and Explosion effect.


Fig. 2 Component overlay for this project.




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# DESIGNER'S NOTEBOOK <br> <br> Remote control is this month's topic, as Don Keighley looks at ways to <br> <br> Remote control is this month's topic, as Don Keighley looks at ways to do things at arm's length - or further. 

 do things at arm's length - or further.}

If you've followed ETI over the years you'll have seen how the trends in remote control have taken place - from simple, single-channel on/off ultrasonic systems through to super-duper luxury systems which allow you to adjust contrast or brilliance on your colour TV without even twitching your leg. The electronics industry being what it is, I don't suppose you will have been surprised by the rapid increases in complexity which were required to produce all the facilities of the later systems. What you may find surprising, however, is that the more complex the systems seem to be, the fewer components are actually inside them! For example, one of the most recent handcontrollers - which offers no less than 32 facilities - uses less than a dozen components (not including key-switches).

This great decrease in the number of components (and hence a new-found ease of system manufacture) is due to medium-scale and large-scale integration. With such integration one IC can do the work of literally hundreds of discrete components, making possible highly complex systems at low prices.

## Distant Data

Quite a number of manufacturers of all nationalities have been jumping onto the remote-control bandwagon - each producing its own varieties. Plessey's 490/922 range provides for remote-control over ultrasonic, infra-red or simple cable links. Pulse position modulation (PPM) is used for data transmission and this gives good noise immunity - even with control distances of about 10 m . A block diagram of the remote-control encoder (SL490) is shown in Fig. 1.


Fig. 1 Block diagram of the SL490 remote control encoder:

The circuit of a basic ultrasonic hand-held transmitter is shown in Fig. 2. A bank of 21 single-pole push-buttons is used in a key-matrix. As each push-button is operated the encoder IC detects closure at a matrix crosspoint and emits a modulated code word of six bits (five data bits and a synchronising bit). The code word is repeated at regular intervals until the push-button is released. A five bit data word allows a range of $2^{5}=32$ different code words - but the matching receiver/decoder IC (SL922) can only decode 21 code words: hence the 21 key matrix.


Fig. 2 Circuit of a 21 -facility remote control transmitter using the SL490 encoder with an ultrasonic transducer.

The ratios of each part of a code word are shown in Fig. 3. Starting with a fixed pulse-length, L , the total length of the code word is dependent on the binary word to be transmitted. The example shown of the binary word 01011 (followed by a synchronising bit of length 18L) takes 54 L to transmit.

The basic pulse length, L, is defined by the values of resistor R1 and capacitor C 1 according to the equation

$$
\mathrm{L}=0.14 \mathrm{C} 1 . \mathrm{R} 1 \text { seconds, }
$$

so for the values shown, the pulse length is approximately 10 mS and so the code word given in the example would take 540 mS to transmit.

Capacitor C 2 and resistor R 2 set the carrier to a frequency given by the equation

$$
f=\frac{1}{C 2 \cdot R 2}
$$

The values shown give a carrier frequency range of $30-50 \mathrm{kHz}$. Most ultrasonic transducers are resonant at frequencies around $38-40 \mathrm{kHz}$ so the component values shown in Fig. 3 are satisfactory.


Fig. 3 Typical output code word of the SL490 encoder. Each part of the code word is defined as a multiple of the basic pulse length $L$. A pulse is formed by a burst of carrier frequency.

The oútputs (pins 2 and 3) from the encoder are in antiphase and can interface directly with the transducer. The range of this circuit would be up to five or six metres. Adding transistors Q1 and Q2 to the circuit (Fig. 4) increases the range to approximately 10 m . A current drain of only 6 uA in standby and 8 mA in transmitting mode means that a standard PP3-sized battery will last for months with normal usage.

If the IC is to be used in an infra-red transmitter, no carrier is strictly required, so components C2 and R2 can be replaced by a single resistor (say 10k) from pin 18 to ground.

## Are You Receiving Me?

The SL922 receiver is shown in block form in Fig. 5. From this you can see that the chip provides:

- a four-bit binary output to control system inputs: such an in-


Fig. 4 The addition of transistors Q1 and Q2 increase the range of the transmitter to about 10 m .


Fig. 5 Block diagram of the SL922 remote control receiver.
put could be a TV channel, audio signal, motor control and so on. With extra decoding circuitry one of a range of up to 16 inputs at a time can be selected

- three analogue outputs (for DC control of volume, tone, colour, brightness etc)
- a low-going pulse output (whenever an input change is selected)
- a toggle output (for muting etc)
- an on/standby output

A typical SL922 receiver circuit is shown in Fig. 6 and from the diagram you'll appreciate how simple it is to use this versatile device. Some form of receiver/preamplifier is necessary to provide positive-going PPM pulses of sufficient amplitude from the received infra-red or ultrasonic signals. Ultrasonic signals must also be demodulated in some way to remove the carrier. Such a receiver/preamplifier could be constructed using op-amp circuitry or a purpose-built IC.


Fig. 6 Circuit diagram of a remote control receiver using the SL922. A preamplifier is required at its input.

## Looking East

Japanese manufacturers didn't take long to get on the scene, of course, and a number of companies produce their own varieties. Nippon Electric Company (NEC) - although more famous for complete equipment such as video cassette recorders and computers - has produced its own range of remote control ICs which is ideal for the electronics designer.

NEC's uPD1986C is a similar device to the SL490, capable of transmitting up to 27 commands. Its block diagram is shown in Fig. 7 and you can see the IC is much simpler than the 490 and


Fig. 7 Block diagram of NEC's remote control encoder - the uPD1986C.


Fig. 8 The code word of the uPD1986C encoder, consisting of three synchronising bits and five data bits.
so is inherently easier to use with respect to circuit design. Standby current is low (less than 1 uA ) so long battery life is obtained too.

A uPD1986C code word consists of eight bits as Fig. 8 shows. The first three bits transmitted (101) are synchronising bits to enable the receiver to detect whether a code word has arrived. A logic 1 bit is represented by a 1.1 mS burst of 38 kHz carrier with a $50 \%$ duty cycle. Logic 0 is represented by carrier absence. Thus, the IC should be usable with either ultrasonic or infra-red transducers without modification.

The remaining five bits of the code word are the data bits of the transmitted remote-control message - corresponding to the key pressed. The code word is repeated every 36 mS until the key is released.

Figure 9 shows the circuit of an 18 -function infra-red transmitter circuit using the uPD1986C. Oscillator frequency is controlled by a 455 kHz crystal to give carrier and code word timing accuracy - a factor which obviously aids long-distance remote-control transmission.


Fig. 9 Circuit of an 18-facility remote control transmitter using NEC's uPD1986C.

Reception and decoding of the transmitted code word is undertaken by the uPD1937C remote-control receiver IC. A block diagram of the IC is shown in Fig. 10, and a typical circuit using it is in Fig. 11. As with the 490/922 chip set, a preamplifier is required to detect and amplify the received low-level signal.

Input selection can be either sequential (ie stepping up or down at a fixed rate) or direct (ie straight to the required input). The output waveform for sequential input-select is shown in Fig. 12a. A sequence step-time of about 0.72 S means that it takes over 12 S to step through from input 1 to input 18. The output waveform for direct input selection is shown in Fig. 12b. The number of pulses in the shift pulse code is one less than the required input number.


Fig. 10 Block diagram of the uPD1937C remote control receiver.


Fig. 11 A typical circuit for a remote control receiver using the uPD1937C. A preamplifier is needed at the input.

(a) SEQUENTIAL INPUT-SELECTION WAVEFORM


Fig. 12 (a) Sequential input select waveform, and (b) direct input select waveform of the uPD1937C remote control receiver.

These waveforms need to be interpreted by decoding circuitry so that the chosen input is switched through. A decoding interface could be easily made with a 4017 or similar counter alternatively NEC manufacture input selectors specifically for the job.

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## HOW IT WORKS

When the device is turned on, Q2 will start to conduct due to current flowing via L1a and R1 into its base. This causes the supply voltage to appear across L1b, which by transformer action increases the voltage (and current) available in Q2's base circuit. This ensures that Q2 will be held on during this part of the cycle. As L1 is an inductance and the voltage across it is by now fairly constant, the current flowing through it will rise linearly. However, the current also flows through R3, causing a voltage to appear across it which is proportional to the instantaneous value of this current. When the voltage across R3 becomes large enough, Q1 is turned on and robs Q2 of its base current. Q2 promptly rurns off and its collector voltage rises sharply. This rise is coupled back to Q1 base via C2 to keep Q1 on during this period. At the same time, the voltage across 11 c will be in a direction such that D3 conducts and charges C3. A short time later Q1 will turn off as it has no base drive once C2 is charged, and Q2 will turn on to start another cycle. D2 provides a discharge path for C2 at the start of a cycle as Q2 turns on.

If when $C 3$ is being charged its voltage exceeds about 500 V , the voltage across L1a will be greater than the supply (due to transformer action with L1c) and D1 will conduct, diverting the excess energy back into C 1 and the battery. The 500 V on C 3 is applied via R5 to the circuit under test. The resulting current then passes through D5 in parallel with M1 and PR1. D4 provides protection if the probes are accidently connected to a live circuit, while D5 protects the meter from excess current and also modifies its response.

R4 in the high voltage section ensures that the output voltage dissipates quickly when the instrument is turned off.


## The Instrument

The project is designed to use a standard PP6 9 V battery and contains a low power DC-DC converter to produce an output of about 500 V . This output is limited so that it does not rise too much even when off-load. The output current is also limited; about 500 uA maximum, even when shortcircuited. Even so IT BITES! - so be careful!

In use the test leads are connected to the circuit under test and when the
button is pressed, the circuit generates a high DC voltage which is applied to the test leads via a 1 M 0 resistor. The resulting current through the insulation is monitored by the meter and displayed as a resistance. When the button is released, the internal capacitor and the circuit under test are discharged fairly rapidly to avoid the risk of shock.

## Construction

The coil is constructed using a


Fig. 1 Circuit diagram of the DVMeg insulation tester.

Neosid potcore. Wind 220 turns of 40 SWG enamelled copper wire on to the former in four layers; start at pin 2 and finish at pin 7. Each layer should be about 55 turns and as this is less than the width of the former, the space each side should be filled with a single layer of insulating tape $2-3 \mathrm{~mm}$ wide. A layer of tape the width of the former should be laid on top of each winding. Next wind on 22 turns of 32 SWG enamelled copper wire starting at pin 3 and finishing on pin 4. Insulate this as before and then wind another four
turns of the same wire starting at pin 5 and finishing at pin 6 .

Construction of the circuit board should pose no problems so long as component polarities are observed. The wires to the test probes should be flexible and well insulated.

The two types of transistor specified for Q2 have different connections, so the TIP31 must be mounted 'upside-down' if used (see diagram).

Installation into the box is also fairly simple. The meter is fitted at one
end of the box lid and the push-button just below it, but a little to one side to allow the battery to fit inside the case. The circuit board is stuck or bolted to the inside of the case lid and the battery leads connected via the switch. The meter leads are also connected, after which the battery may be connected and fixed into place with sticky pads or wedged with foam.

## PS <br> It also tests neons!

Fig. 2 Component overlay and component pinouts for the DVMeg.


Fig. 3 A suitable meter scale.

## BUYLINES

The pot core used to wind the inductor for this project is available from Neosid Small Orders, PO Box 86, Welwyn Garden City, Herts AL7 1AS; quote ref. 29-835-41 when ordering. West Hyde supply the case we used; for people who want to use substitutes the size is $150 \times 80 \times$ 55 mm . The PCB can be obtained using the order form on page 45.


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# PRIZE CROSS WOR 

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Time once again for our every-alternate-month stretching of your brains - the ETI Prize Crossword. Many of you obviously found Crossword No. 2 more difficult than the first one, since few readers were able to complete it correctly; but of those that did, the following three were picked out as our prizewinners:
FIRST PRIZE: H.R.W. Thurlow, Gravesend SECOND PRIZE: A.M. Tucker, Dorchester THIRD PRIZE: Nicholas Brasier, Frimley Green Congratulations to them, and good luck to everyone who enters for Crossword No. 3; hopefully we've made it less obscure!

## ACROSS

## LAST MONTH'S SOLUTION



1. Switching device from $R X$ to $T X$. (8)
2. Rotating member in the motor. (8)
3. 'Son of Crystal'. (7)
4. Roman numeral shapes up for a trip. (7)
5. Short magazine has centre spread between I and E to get picture. (5)
6. The initial effect on transistor by farmer's land. (3)
7. The right place for electronic discharge. (3)
8. Backwards or forwards it indicates the same. (5)
9. Something more than an ordinary BC182. $(1,4)$
10. Signal selector. (5)
11. West, central in twisted rope. (5)
12. Fix the result - Good Buddy! (3)
13. Timely happening. (5)
14. Re-organise magnetic formation. (5)
15. Airborne beacon electronic test set - may help initially. (5)
16. Unit of luminous flux. (5)
17. Broken cart - ideal vision. (3)
18. Perhaps an about turn would change a short sagging economy ready for inflation. (3)
19. A symbol no good ohm should be without! (5)
20. Reagan changes tune about radioactive weapon. (7)
21. Used by Ken Dodd to trigger laughter? (7)
22. Port side, one of a disc jockey's pair. $(4,4)$
23. Rest period for computers. $(4,4)$

## DOWN

1. 10-1 your ears may ring at these excessive levels. (8)
2. Precisely the right approach for golfers when using radar. (3)
3. Go east, go north, then round and round. (8)
4. Found in radio cabin and Laker Airlines! (8)
5. The French are crazy about celestial visitors. (3)
6. Computer's credit card - not this side of the channel. (8)
7. Twice as good as stereo. (4)
8. Fortune finder? $(5,8)$
9. Backward thinking? $(8,5)$
10. Vertical reference. $(1,4)$
11. Trim Lane roundabout for meeting point. (8)
12. Hurt coin produces second wave attack. (8)
13. Floating input may be lonely. (8)
14. Mixed hen cargo - great top-up! $(2,6)$
15. Highest point of AC cycle. (4)
16. TV or not TV, that is the answer. (3)
17. Large Scale Integration. (3)


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| 7490 | 4017 | 50p |  |  | ${ }^{\text {LS } 157}$ | 57p | ( ${ }^{16 \text { poin }}$ | ${ }_{16 \mathrm{p}}^{109}$ | 3.75" $\times 3.75{ }^{\prime \prime}$ | 83 p |
| 93 | 4020 | ${ }^{600}$ | Ls02 | 15p | LS160 | 54P | 24 pin | 20 P | 5. $\times 3.75$ | 95 p |
| (107 | ${ }_{4024}^{4023}$ | ${ }^{190}$ | ${ }_{\text {LSO }}^{\text {ise }}$ | 15 c 150 | ${ }_{\text {LS164 }}$ | 54p | ${ }^{280 \mathrm{pin}}$ | ${ }_{32 \mathrm{p}}^{24 \mathrm{p}}$ | 3.75" ${ }^{4} \times 17$. | 326p $\mathbf{4 2 6 p}$ |


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# SLOT CAR CONTROLIER 

# Some spare cash and an idle Saturday afternoon led into the labyrinthine world of slot cars. The basic slot car set is so basic that we decided to improve it, and this project is the result. Design and development by Jonathan Scott. 

Well, let's not beat about the bush. Slot cars are fun. The genesis of this project was the purchase of a cheap set and the realisation that there was much room for improvement in the whole thing, especially the 'electronics'. Since then, we have built several controllers, purchased an alarming length of track, bought and modified many controllers and cars, and generally had a load of fun!

In the course of this research, use has been made of calculators, programmable calculators, desktop computers, plotters, engineering degrees, physics degrees, computer science degrees, a mound of components, a lot of paper and a hell of a lot of electricity - so be warned that one can get pretty involved. Closet racers, prepare for exposure!

## The Ultimate Controller

If you're after something really exciting, then this is it - but it's not a project for beginners. The controller gives independent supplies for each lane and can operate in voltage and current modes; it can handle a wide range of maximum torques on sets of 4 V 5 to 12 V rating.

For superior performance, the controller has several 'extras'. First, there is fuel tank simulation; this means that the control box has a meter which represents fuel in the car. A button 'refuels' the car, provided it is stationary. When it has petrol, you can go again. As the petrol is used up the car gets more acceleration, corresponding to the 'reduction in weight'. The degree of the effect is presettable by a resistor (R7 - or R107 for the second car). It is rather exaggerated with the value given, but this is more fun. Of course, if you run out of fuel, the car slows down and finally coughs to a stop.

Next, the project offers controlled
overshoot. If the output momentarily exceeds the level that your hand controller commands, the car responds more 'snappily'. This accelerates it a bit harder at first, corresponding to 'dropping the clutch', and brakes hard when it is slowing down corresponding to hard braking. You can even lock up, if you are too hasty!

The controller also informs you if it is folding, such as when the track is short-circuited. In the current mode, it warns of open circuit as well. It does not load the hand controller rheostats, as they do not carry the car current. (In some sets the controller handsets get very warm.) The controller is, of course, short-circuit protected.

The two modes, current and voltage, each offer their own advantages. Current mode gives torque proportional to control depression, as torque is proportional to current. It has slower take-off and generally sloppier (though perhaps more realistic) operation. It is also more immune to bad contact in the track and brushes, if you are having trouble in that direction. Voltage mode, which we prefer, gives a very tight control with a snappy response from the car; perhaps less realistic, but more fun. It seems to demand more from the drivers, though performance is considerably superior.

## Construction

Construction of this unit is fairly flexible and will depend somewhat upon how you plan to house the unit. It is advisable to use a fairly spacious housing as this demands less careful layout and allows easy access for adjustment or debugging. The only requirement for the case is that if you are using our PCB the meters must be spaced horizontally by the required amount, as the board mounts on the meter terminals.

First step in the construction is to drill the case and panel. Note that the power supply transformer and rectifierfilter components are not included on the PCB and are mounted on the case in a convenient position. You will have to position these components so that they do not foul any other components, and drill the case to suit. The power supply components are mounted off-board for a number of reasons. First, they are bulky and would add seriously to the PCB size and the space needed for it, and second, it means that a set of higher current capacity diodes and a higher VA-rating transformer may be used for powering more than two lanes. A 40 VA
transformer will power two lanes, a 60 VA will power up to four lanes.

We found it convenient to mount


The slot car controller board before interwiring.


Fig. 1 Circuit diagram of one channel of the controller.


Fig. 2 Circuit diagram of the power supply.
the mains supply terminating block, cable clamp (or clamp-grommet), output terminals and presettable pots ( $\mathrm{R} V 1, \mathrm{RV101)}$ ) on the rear panel of our box. We used ordinary potentiometers for RV1 and RV101, rather than preset types, cut the shafts short and cut a slot in the end of the shafts. To avoid fouling other components, mount the pots so that they are below the height of the transformer

Next, prepare the front panel. Drill it first, locating the meter holes carefully as the PCB determines their spacing. For panel marking we used rub-down lettering (such as Letraset, etc) put directly on the panel after cleaning it.

Apply a spray-on lacquer to protect the panel markings. With this job finished, fit the meters, LEDs, and so on. Finally, drill the mounting holes for the power transistors, which are mounted off the board. These dissipate little heat so they merely need mechanical support.

The next step is to assemble the components to the PCB. As there are quite a few flying leads, it may pay to use pins for the termination of these to the board Pay attention to all the usual details - orientation of tantalum and electrolytic capacitors, orientation of semiconductors, etc. Choose the components in Table 1 to suit your requirements, according to the
instructions given with the Table. When all the components are soldered in place, fit the flying leads to the LEDs and push-buttons which are mounted on the front panel, along with the meters. These can be secured and the PCB bolted to the meter terminals before the leads to the main case are fitted. Be sure that all flying leads are long enough to allow the box to be fitted together and dismantled without straining the connections. Long leads can be kept neatly 'loomed' with plastic sleeving slipped over a bunch before one group of ends is terminated.

Assemble the transformer, power supply components and potentiometers in the case next and wire them up. Take particular care with the mains wiring. The rectifier components are supported on a tagstrip and we'll leave the wiring details to you for this one.

The final step before testing is to modify the hand-held controllers from rheostats to true potentiometers. Open up the case of a controller and you will find that it consists of a short coil of resistance wire wound on some sort of former, with a wiper contact which moves along the coil according to how far the thumb or finger control is depressed. When fully released, the wiper rests in a position where it does not touch the coil. There will be two wires coming from the hand controller - one leading to the wiper and one from an end of the resistance wire. It is

## HOW IT WORKS

The unil comprises a power supply, a control section (involving IC1), a driver circuit (involving Q4, 5 and 6) and associated components), an overload protection and warning circuit (Q2, 3 etc), an 'electronic fuel tank' (Q1 plus IC2 and associated components) and a 'refuel' circuit (Q7 etc).

The circuit has two modes of operation - voltage and current. The mode to be employed is selected by means of a link on the PCB. In the voltage mode, the hand controller sets the voltage delivered to the track (and thus the slot car's motor). In the current mode the hand controller sets the current delivered to the car's motor via the track. In either mode, a potentiometer ( $R$ V1) sets the maximum value of the voltage or the current.
POWER SUPPLY
Transformer 11 has two 15 V (RMS) secondaries, connected in series. There are two rectifier circuits - one to provide a positive supply rail, the other to provide a negative supply rail. The joining of the two secondaries provides a 0 V connection.

Diodes D5-D6 and capacitor C5 provide a nominal +21 V supply rail $\left(+V_{p}\right)$ while D7-D8 and C6 provide a nominal $-21 \vee$ supply rail $\left(V_{N}\right)$. From these two rails +12 V and -12 V regulated rails are derived. The +12 V rail is achieved by IC3, a threeterminal positive supply regulator (a 7812 or 78L12). This rail is used as a reference for the hand controller and metering crcuit. Capacitor CB ensures high frequency stability for the threeterminal regulator and acts as a supply rail bypass. The -12 V rail is derived by a simple zener circuit involving R26 and ZD2. C9 is a supply rail bypass. The negative rail is limited to 12 V so that the maximum supply voltage limitation of the op-amps, which is about 36 V , is not exceeded.

## CONTROL SECTION

This centres on IC1. A certain current (which we will discuss in detail a little later) is passed through the hand controller resistance. This develops about 200 mV drop across it. Thus, when the hand controller is operated, a voltage ranging between 0 and 200 mV is applied to pin 3 of IC1, the précise voltage depending on how far the 'driver' has depressed the controller lever. C3 smoothes out any variations - many hand controllers have momentary loss of contact between the wiper and the resistance as the wiper traverses the resistance element. You may need to vary the value of C3 according to how coarse the resistance variation happens to be in your controller. For the inexpensive controllers which are really quite adequate despite the coarse variation they provide - a value of 470nF to 1uF is suitable.

Now, IC1 attempts to drive its output (pin 6) in such a fashion as to induce the same voltage on its inverting input (pin 2 ) as is on its non-inverting input (pin 3).

In the voltage mode, pin 2 of IC1 is connected via RV1, C2 and associated components to the positive track terminal so that the position of the wiper on the hand control resistance sets the output voltage. In the current mode, pin 2 of IC 1 is connected to the end
of the 'current sense' resistor (R16) so that current is defined by the position of the wiper on the hand controller resistance.

In either mode, RV1 - which is in series with the negative feedback path - in conjunction with R1 sets the maximum vollage or current delivered to the car's motor via the track. C2 induces some 'overshoot' in the feedback which enhances acceleration and braking according to controller movement.
DRIVER
The driver circuit comprises Q4, Q5 and Q6 plus R25. Its function is merely to amplify the current delivered from the outpul of IC1.

Transistors Q4 and Q5 are connected as a Darlington pair which provides considerable current gain (the beta of Q5 is multiplied by the beta of Q4). The output of IC1 (pin 6) swings positive during acceleration (depressing the hand controller lever) and Q45 amplify the current, the emitter of Q5 being connected to the track positive terminal. Q6 is reverse biased during this time. During braking, pin 6 of IC1 can go negative (particularly if you 'drop' the hand controller lever). This reverses the voltage delivered to the track or reverses the current flow (depending on which mode you're employing). When this occurs, Q4 and Q5 are reverse biased and Q 6 is forward biased - and it amplifies the negative excursions from pin 6 of IC1.

The function of R25 is to protect Q6 against momentary current overload.

## PROTECTION

The protection circuit involves $\mathbf{Q} 2$, Q3 and associated components. If the voltage output to the track exceeds about 13 V, ZD1 and D4 conduct, forward biasing the base of Q3. When Q3 turns on, it draws collector current via LED2 and R21. LED2 lights, providing warning of a fault. If the output current exceeds about 1A5 the current through R16 (which is in series to the supply to the track) induces a voltage drop across it of about 0 V 7 or so and this forward biases the base of Q2 via R17 and R18. Q2 thus turns on and it draws collector current via D2, R21 and LED2. However, the collector voltage of Q 2 will be around a few hundred millivolts and the output of IC1 will be shunted to the 0 V rail via D3 and the collector-emitter junction of Q2.

Thus, you receive a warning of supply overload and the supply, track etc., is protected against overcurrent damage.

## FUEL TANK

The 'fuel tank' is simulated by IC2 and associated components. This op-amp is connected as an inlegrator. A 'full' lank corresponds to 0 V on the output of IC2, an 'empty' tank to about 12 V . As current flows through the load (car motor), and hence via R16, a voltage is dropped across R16. This voltage is integrated by IC2 which has an RC network (R13-C4) in the feedback loop. As more load current is drawn, pin 6 of IC2 rises towards 12 V .

The meter, M1, indicates the output voltage of IC2 and is marked like a fuel gauge. While the fuel tank is full or partially full, the current through M1 flows via the base of Q1,
forward biasing it. Thus, Q1 is held on while this current flows. The collector current of Q1 flows via LED1 (the hand controller and associated resistors). LED1 lights, indicating you have fuel in the tank. When the fuel 'runs out', pin 6 of IC2 is at 12 V and no current flows through M1; thus the base of Q1 receives no bias and it turns off. LED1 extinguishes at this stage and no voltage is delivered to the hand controller. IC1 interprets this as if you have the controller sel to the rest or off position and no power is supplied to the track. Your car stops.

The 'capacity' of the fuel tank is defined by the values of C4 and R13. The values shown give a 'full tank' of about 60 amp -seconds which corresponds to about 30 rapid laps of a $21 / 2$ metre long track in $1 / 64$ th scale. The values of C4 and R13 may be varied to suit your taste, as indicated in the table on page 33 .

While there is fuel, LED1 is on and its terminal voltage is about 1 V 7 . This voltage permits about 10 mA to flow through the resistance of the hand controller via R5. (Recall we have yet to see what its current is.) In addition, R7 permits some current to flow into the controller - generally between 0 and 5 mA - from pin 6 of IC2. This current increases as fuel is 'used up', corresponding to the car getting lighter, and you get more acceleration at any particular hand controller setting as you 'use up' fuel. R7 defines how much more acceleration is obtained when the car is 'lighter'

When the fuel runs out and Q1 turns off, the current delivered through $\mathbf{R} 5$ to the hand controller plummets and only the 5 mA flowing via $R 7$ is available. This gives a 'soft' end, allowing you to limp to the pits - if you aren't too far away on the track.

The parallel combination of R8 and the hand controller should be around 15 ohms. If .your controller has a high resistance, or you want to substitute a 1 k 0 wirewound pot, for example, R8 should be derived from the following formula:

## $\mathbf{R 8}=\mathbf{R C O N T R O L L E R} \times 15$ <br> RController - 15

## REFUELCIRCUIT

'Refuelling' is effected by PB1 and Q7. When the car motor is not drawing power, the output of IC1 is low (less than one volt) and thus Q7, which derives its base bias from pin 6 of IC1, is off. Pressing PB1 connects R12 to the +12 V rail via R 24 and IC2 will discharge C4. The outpul of IC2 will drop to 0 V (which is the 'tank full' condition). Q1 will turn on again and current will be supplied to the hand controller circuit. When you power the car again, the voltage on pin 6 of IC1 will rise, the base of $Q^{7}$ will be biased on and its collector will draw current via R24. Thus, if you Iry to 'top up' while the car is in motion, R12 will be virtually connected to the 0 V rail via the collectoremitter junction of Q7 and you won't be able to drive the output of IC2 low. In addition if you attempt to drive the car while refuelling, the refuelling action will be stopped by the same means.
necessary to have a third contact, connected to the other end of the coil (the end without a connection). Remove the existing wires (some of these have considerable resistance themselves) and fit the two new wires, then the third. These run to the controller unit. Make sure you have plenty of length to play with. Now re-
assemble the hand controller, being careful to tie off the wires in the same way the original two were secured. You should now be ready for a test run.

## Test Run

Make up a simple circle of track. On powering up, the car should work
to some degree. If not, stop and recheck. Once it works it is necessary to adjust the presets and so forth. PR1/101 should be adjusted to minimise 'fuel tank' circuit drift in the absence of power being delivered. (These are the integrator offset adjustments.) At this stage it is probably worth assembling the unit


Fig. 3 Component overlay and interwiring diagram for the ETI Slot Car Controller. Note that we mounted three of the transistors for each channel on the case itself, and that some of the power supply components are mounted separately on a piece of tagstrip. The board itself is fastened to the front panel by means of the meters, which are bolted directly to the PCB (you can see the bolt heads in the photo on page 79). A choice of mounting pads is provided to allow for different meter dimensions - most 2" panel meters should fit, but check before buying.

TABLE 1

| COMPONENT | NOMINAL VALUE | FUNCTION | HOW TO VARY IT |
| :---: | :---: | :---: | :---: |
| C4 (C104) | 10u | Sets fuel tank capacity. | Increase its value to increase fuel tank capacity. Range: 10 to 47 u . |
| R13 (R113) | 47k | Sets fuel tank capacity, along with C4. | Increase its value to increase fuel tank capacity. Range: 10k to 100 k |
| $\begin{aligned} & \text { R10,11 (R110 } \\ & 111) \end{aligned}$ | 120k in parallel with 12k | Calibrates M1 for full scale deflection at 'full tank' status; allows other meter FSD values to be used. | Reduce $\mathbf{R} 10$ to increase reading. Choose R10/R11 to give value according to $11.4 / \mathrm{I}_{\mathrm{FSD}}$. This should not need much adjustment if a 1 mA meter is used. |
| R7 (R107) | 2k7 | Sets the variation of engine power with remaining fraction of fuel. | Reducing $\mathrm{R}^{7}$ gives a greater gain in power as the fuel runs out. Range: 2k2 to 22k. |
| R8 (R108) | 100R | Sets the effective controller resistance to about 15 ohms. | Choose R108 so that R8 in parallel with the controller resistance gives a combined resistance of 15 ohms. |

PARTS LIST

| Resistors (all $1 / 4 \mathrm{~W}, 5 \%$ except where stated) |  |
| :---: | :---: |
| R1 | 1k0 (link A), 39k (link B) |
| R2 | 10k |
| R3,20,21 | 1k0 |
| R4 | 82k |
| R5 | 470R |
| R6 | 150R |
| R7 | 2k7* |
| R8 | 100R* |
| R9,24 | 22k |
| R10 | 120k* |
| R11 | 12k |
| R12 | 270k |
| R13 | 47k* |
| R14 | 27k |
| R15 | 100k |
| R16 | OR47, 1 W |
| R17 | 100R |
| R18 | 390R |
| R19 | 2k7 |
| R22 | 3k9 |
| R23 | 3k3 |
| R25 | 10R, 2 W |
| R26 | 270R |
| Potentiometers |  |
| RV1 | 100k linear |
| PR1 | 10k miniature vertical preset |
| Capacitors |  |
| C1 | 1 n 0 ceramic |
| C2 | 3 u 310 V tantalum |
| C3 | 100n polyester |
| C4 | 10 u 16 V tantalum |
| C5,6 | 2200u 25 V axial electrolytic |
| C7 | 470u 25 V axial electrolytic |
| C8,9 | 10u 16 V tantalum |


| Semicondu ctors |  |
| :--- | :--- |
| IC1 | LM301 |
| IC2 | CA3140 |
| IC3 | 78L12 or 7812 |
| Q1 | BC179 or BC559 |
| Q2,3 | BC107 or BC547 |
| Q4 | BD139 |
| Q5 | 2N3055 |
| Q6 | TIP32 |
| Q7 | BC109 or BC549 |
| D1,3,9 | 1N4001 |
| D2,4 | 1N914 |
| D5-8 | 1N5404 |
| ZD1 | $11 \mathrm{~V}, 1 \mathrm{~W} 3$ zener |
| ZD2 | $12 \mathrm{~V}, 1 \mathrm{~W} 3$ zener |
| LED1 | 3 mm red LED |
| LED2 | 3 mm yellow LED |

## Miscellaneous

T1 0-15-0-15 40 VA transformer (or SW1 $\quad$ DPDT mains-rated toggle switch PB1 momentary action push-button M1 $\quad 1 \mathrm{~mA}$ FSD meter
PCB (see Buylines); case to suit (see Buylines); tagstrip; terminal block; mains cord and plug; clamp grommet; mounting hardware etc.

## NOTE

The controller circuit is duplicated for the second track, so you will require two of each component with the exception of R26, C5-9, IC3, D5-9, ZD2, SW1 and T1. Components for the second controller are designated R101, C101, D101 etc.
*Components marked with an asterisk may need alteration to suit your particular require ment (see text).
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 mention E.T.I. when replying to all advertsand giving it a serious workout. You may find that you want to increase the fuel tank capacity ( C 4 ), change from one mode to the other (links A and B) or that the control is rough or jittery. If this latter is the case, then your controller is probably one with relatively few turns of resistance wire. This is causing sharp changes in level, to which the electronics respond with excessive overshoot. The cure is to increase C3/103 to, say, 1 uF . This is especially prevalent with the cheap,

6 V operated sets. After you have had a while in the seat, remove the front panel and alter the appropriate components (marked with an asterisk) in order to produce the effects desired. To find out what these are, consult Table 1.

A note should be included on the correct adjustment of the maximum torque presets, RV1/101. This is much a matter of preference. They should be adjusted so that the car does not get ridiculous amounts of power just prior to running out of fuel, but so that the car can just be crashed on full power with a full tank. It is probably also a good idea to set the two channels alike with a multimeter to ensure fairness. (Be sure to have equal amounts of fuel when doing this adjustment!)

ETI


#### Abstract

\section*{BUYLINES}

There should be no problems obtaining any of the components used in this project. Note, however, that the PCB is laid out to accept TO5-packaged CA3140s - not the Dil type - so make sure you buy the correct ones. The case we used is one from the Vero range, order no. 65-2523E, but any similar-sized one will do (about $220 \times 156 \mathrm{~mm}$ ). The PCB can be obtained from us at a cost of $£ 3.72$ by filling in the order form on page 45. $\qquad$

\section*{}號


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# READ/WRITE 

Letters for this page should be addressed to Read/ $\overline{\text { Write }}$ at our Charing Cross Road address.

Dear Sir,
I am afraid Vivian Capel, in his User's Guide to Microphones in the Feb. 82 issue, is not quite correct in his otherwise excellent article when he states that the resonance of a ribbon microphone is "high, near or beyond the normal range." In fact, the resonance in most ribbon mikes is in the region of 35 to 45 Hz and is usually damped by the use of resistive screens placed close to the ribbon.

One of the reasons why the ribbon type exhibits a warm, natural tone is that the resonance is low down and not at, or above, the high frequency range as is the case with capacitor types, some of which display a very hard tone especially with stringed instruments.
The upper end of the ribbon's frequency response trails away gradually and this gives the warm tone when compared with the capacitor.

For further information I would refer Mr. Capel to the excellent $B B C$ Engineering Monograph No. 4 of 1955, which gives a lot of detail on the design of ribbon microphones.

## Yours faithfully,

Cyril Cott,
Southall.

Dear Sir,


Did you know ETI has improved a lot lately. This is the first time I've ever written to a mag, I hope you will print this in your next ETI and answer my following questions:-

1) The Pest Control Feb 82 - can it give out more power, because I tried it near a cat and it didn't do anything.
2) The Computer Expansion System can it be used on the VIC 20 and what is the difference between RAM and ROM. Yours faithfully,
Sukkin Pang,
Harlow.
PS. Keep up the good work.
PPS. Happy birthday.
3) Stepping up the supply voltage would give more power but isn't recommended because the components would complain! It
shouldn't really be necessary to get more power - we had a phone call from a large shopping centre whose automatic doors were letting in all the local cats and dogs until they fitted one with an Allez-Cat, whereupon the animals fled. They now intend using Allez-Cats on all the doors; are you sure the cat in question isn't deaf?
4) The computer expansion will work with the VIC 20; we'll provide details of how to do this in a later issue. RAM is random access memory which is easy to alter but forgets everything when the power supply is disconnected. ROM is Read-Only Memory which retains information even when the computer is turned off, but in the form the hobbyist normally uses (EPROM) data can only be changed by erasing the chip under ultraviolet light and then reprogramming it using a special series of pulses.

## Dear Sir,

I recently purchased an amplifier kit from Hart Electronic Kits Ltd, Oswestry. When built, it worked the first time. The quality of parts was excellent and the packing was very good. Even the finish was first-class. However, there was one small omission - no base for the relay.

Upon phoning them (a lady was very helptul) the missing part arrived by return of post. So 1 am very pleased and will use this firm again because they obviously wish to provide a good service and should be encouraged. If you publish these facts in your Audiophile article perhaps other readers may benefit.

Oh yes, I almost forgot, I enioy your magazine; especially the way things are explained, because it's hard to teach old dogs like me new tricks. Please keep Audiophile going - it's a great little article. I have purchased the Coral MC88F from your suggestion and, again, am very pleased with it anyone should be at that price.

Cheerio now and watch that blood pressure!

Yours faithfully,
L.T. Bowler,

Cardigan.

My thanks for your medical meditation. My blood pressure is firmly under contyol at most times in life (the
exceptions are those things worth living for. .). It is a welcome change to hear tales of wonder rather than of woe. Any other readers who can recount the good and chivalrous deeds of companies - let's hear about it! We could all do with good news these days.

Dear Sir,
With reference to your article Pickup Amp Design, Jan. 82.

I have a problem in determining the correct values of $C$ and $R$ to accommodate the pre- 1955 standards for 78 rpm records where the bass turnover frequency can be between 300 Hz to 450 Hz depending on the manufacturer. Also the 50 uS time constant gives a 3 dB point at a different frequency from post-RIAA.

I would be grateful if your next article in the series could contain a figure to indicate the values of $C$ and $R$ for the time constants 3180 uS, 450 uS and 50 uS . Is it also desirable for the reasons mentioned in the article to attenuate the bass as per new RIAA?

The text and Fig. 2 of the article both indicate the old RIAA bass time constant to be 3150 uS, should this be 3180 us?

Can you please pass this letter to Mr. Tilbrook.
With thanks,
Yours faithfully,
R.F. Butson,

Cardiff.
Your comments will be passed on to Dave Tilbrook; unfortunately he writes for our Australian edition, so it may be a while before we get a reply!

## Dear Sir,

Re: Computer Controlled Live Music, Feb 82.

In recent months I have begun to see many cases of micro for micro's sake but I believe that some of the ideas in this article have blown the subject up out of all proportion to its usefulness and have gone about using it in the wrong way.

I am speaking from the point of view of someone who is working towards the use of computer (don't forget common-or-garden digital) control for both live music and studio recording and I have talked with several professional and semi-professional engineers and musicians about the subject.

Some of the points that have arisen are these:

1) Any small band (such as those at which this article was aimed) not using a

PA and/or mixer system, does so because they can't afford such things, or they want to keep their set-up simple for the benefit of the music. No one in this situation is going to want to buy a $£ 500$ micro and display to control their two or three amps.
2) I have never seen a band without a mixer, who are content to leave their amp and effect controls alone throughout a set, because they 'couldn't hear themselves' or they 'thought it could be that little bit better'. The point that arises here is that either the black box takes over from the amp controls completely, which no self-respecting musician would allow, or the controls work in parallel to the computer. In this case any itchy fingers on the controls would render any future changes by the computer totally wrong.
3) Any computer control system for live music must be continously monitored and updated. No two live performances are exactly the same, especially since any mix in an empty hall is generally wrong once you've got an audience. Such continuous control can only come from an engineer sitting at some sort of mixing console. For instance, what would happen if half way through your ZX81-controlled set, someone knocked the 16 K RAM pack? The best place for any computer in a PA chain is therefore within the arm span of some single completely trustworthy person.

A live music system with computer control can therefore be reduced to a computerised, or even just digital, mixing desk. EQ changes during a performance are usually few and far between and stereo panning at high sound levels is too subtle for constant use and so can be left to the human side of the band (if there is one left). Your hyper-expensive-computerised-personal-amplifier system is thus reduced to a bank of voltage controlled amplifiers linked up to a small memory. You don't even have to spend a thousand quid on a Roland Computamix to do it either. Such a system comes into its own if it is part of the mixing desk in the first place.

No band or individual musician should be given the idea that if you don't tag along with computers you'll get left behind. It is the exception rather than the rule that any home brewed micro-systems will ever improve your music or even save time. If you want to use a computer for live music, first check your motives and reasons and if you're still certain, then start with it cheap, simple and easy to operate. It saves wasted time and money and the loss won't be so great as to put you off computers for good.
Yours,
Dave Pallant,
London.


## Dear Audiophile,

Two years ago I built ETI's Audiophile 4000 system, the power amp is, I think, a very good design and the sound quality, for which I can vouch, is of the best quality.

I have used this power amp with many preamps including the one designed for the 4000, although in the case of the 4000 preamp I did a mod by deleting the filters.

Last year when System A was published I built the preamp with MM input because I thought it offered improvements over the 4000 preamp and also there are no tone controls.

I didn't really expect such marked improvement as I have got from this combination of ' 4000 power amp and System A preamp'. The improvement in sound detail, separation, clarity and bass response is really satisfying.

I wonder if any other readers have tried this combination.

I am about to change my tuner, it is an old Sony STR-6040 tuner amp from the early seventies but I only use the tuner section, it has been the most reliable piece of equipment I've ever owned. I am interested in a Sony ST-I75 tuner which I have seen mentioned in two of your Audiophile articles. I wonder if you would comment on its pros and cons.

My present system consists of:
System 'A' preamp with MM input
Audiophile 4000 power amp
Thorens TD160S SME Mk III with
Technic EPC 205 Mk III
Technics RS615 Cassette Deck
Rogers Studio 1 Speakers
I have noticed that you never published a MOSFET Power Amp, I wonder why?

In conclusion I am a very satisfied reader of your very good magazine.

I remain,
Yours faithfully,
I.R. Worrell,

Wood Creen.
The ST-I75 is a good tuner. A full review appeared in ETI June '81. It is capable of very clean reproduction indeed and should serve you well. (listen to the Revox too if you get the chance.) Against the Sony is the fact that it could handle large scale works a little better than it does, having a tendency to sound confused. A minor blemish though.

What do you mean, we haven't published a MOSFET amp? Of course we have. As punishment for not knowing instantly which issue it was in I'm not going to tell you! look to your indexes, sir.

## Dear Mr Harris,

Your excellent magazine's latest article on 'Buying Mail Order' has prompted me to contact you.

1 am currently trying to start my own business and most of the suppliers I am using at the moment are companies I have used in the past through my hobby.

The different ways in which companies treat their customers is almost unbelievable, and I think that if a magazine could run a 'league table' of particularly good and terribly bad suppliers it would be a useful guide to prospective customers and would also encourage suppliers to keep on their toes.

Example 1: I am in the process of buying a VDU for my home computer which I am adapting for business use. I wrote to two of your regular advertisers that claim to have VDUs for sale. In both cases I have asked for technical data, warranty conditions if any and delivery time and price.

Neither company has bothered to reply to my enquiry! - but they are both still advertising VDUs. I wonder why, if they will not sell them?

Example 2: I was recently stuck for some resistors. I had purchased some instrument kits and two kits were short of odd components. I telephoned to see if they could help out. They quoted me a fairly good price and said that they would despatch ASAP.

I couldn't believe it when the Postman handed me a 'liffy Bag' the following morning with all my resistors. Quite something when I tell you I hadn't rung until 4 pm the previous day and my mail arrived at 7.15 am !

## One final point.

Some components that are used in vast quantities are rarely ever advertised, eg insulating materials, hardware such as screws, nuts and washers, grommets, etc and, most of all, wires and cables.

It doesn't matter how complex your PCB might be, it has to connect to the outside world sometime. But if nobody advertises wire, how do we do it?
Yours faithfully,
C.R. Tattersall M.Inst.E, Haslingden.

## If we did as you suggest and started

 publishing blacklists, the libel suits would start flying around faster than insults in the House of Commons!
## Dear ETI,

Thank you for at last introducing the READ/ WRITE forum.
Yours faithfully,
D.P. Allen,

Wembley.

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# SOUND EFFECTS 4 

# Our fourth and final sound effect using the SN76488 should come in handy for any amateur drama group contemplating Agatha Christie it's a gunshot. Design by Phil Wait. 

This unit is quite straightforward. The Noise Generator blocks in the IC are employed to produce a suitable sound, which is heard for about a fifth of a second, dying away rapidly. The effect is triggered (pardon the pun) by the push-button. Only half a dozen components are required apart from the IC! With care, patience and a little juggling, the unit could be fitted inside a toy plastic gun by simply soldering the components between the IC pins. You would have to obtain a tiny loudspeaker, headphone unit or rocking armature insert for a speaker - whatever will fit in the gun assembly.

## Supply Bypassing

A short word on this subject may prevent difficulties in some cases. In general, we found that the power supply rail doesn't really need bypassing. However, provision has been made on the PCB for the inclusion of a bypass capacitor. This is located near the battery positive lead input on the board, which connects to pin 12 of the IC. Have a look at the component overlay for the Bomb Drop and Explosion unit, in last month's ETI. Locate C6, a 10 nF polyester. This is the
supply bypass. A capacitor having any value between 10 nF and 10 uF , and which will fit on the board, will do the job.

That's it! Have fun with your Sound Effacts. We're sure that, with a

## HOW IT WORKS

A gunshot is simulated by producing a burst of noise that decays very quickly. This unit employs the Noise Generator, Noise Filter, One Shot, Mixer and Envelope Generator to generate the required sound.

The Mixer select pin (25) and the Envelope select pin (28) are both held high ( +5 V ), selecting the One Shot output function from the Mixer. When the push-button, PB1, is pressed it puts a high on pin 9 and the Systems Inhibit block triggers the One Shot and activates the Envelope Generator. For the duration of the One Shot period, the modified noise from the Noise Generator/Filter is passed through the Mixer and Envelope Generator and then to the audio output stages.

The One Shot period, determined by R1 and C1, is quite short (about $1 / 5$ second) and the decay period of the Envelope Generator a little longer. Audio output is coupled to the speaker via the 100 uF DC blocking capacitor, C4.
little ingenuity and experimentation, you'll be able to devise a few effects of your own. (If you do, we'd like to hear from you and will pay for any items published.)

Fig. 1 Circuit diagram for the gunshot effect.


Fig. 2 Component overlay for the gunshot, using the same PCB as all the other sound effect projects. Don't forget to insert the link on the board.

## WHAT'S ON NEXT?

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## HE Reverb

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