

TRANSCENDENT POLYSYNTH

By brilliant design work and the use of high technology components the Polysynth brings to the reach of the home constructor a machine whose versatility and range of sounds is matched only by ready built equipment costing thousands of pounds. Designed by synthesizer expert Tim Orr and being featured in this issue of Electronics Today International, this latest addition to the famous Transcendert family is a 4 octave (transposable over 7½ octaves) polyphonic synthesizer with internally up to 4 voices making it possible to play simultaneously up to 4 notes. Whereas conventional synthesizers handle only one at a time.

The basic instrument is supplied with 1 voice and up to 3 more may be plugged in. A further 4 voices may be added by connecting to an expander unit, the metalwork and woodwork of which is designed for side by side matching with the main instrument. Each voice is a

complete synthesizer in itself with 2 VCOs, 2 ADSRs, a VCA and a VCF (requiring only control voltages and a power supply, the voice boards are also suitable for modular systems). One of these voices is automatically allocated to a key as it is operated. There are separate tuning controls for each VCO of each voice. All other controls are common to all the voices for ease of control and to ensure consistency between the voices.

Although using very advanced electronics the kit is mechanically very simple with minimal wiring, most of which is with ribbon cable connectors. All controls are PCB mounted and the voice boards fit with PCB mounted plugs and sockets. The kit includes fully finished metalwork, solid teak cabinet, professional quality components (resistors 2% metal oxide or metal film of 0.5% and 0.1%), nuts, bolts, etc.

EXPANDABLE
POLYPHONIC
SYNTHESIZER

POWERTRAN

TRANSCENDENT POLYSYNTH

COMPLETE KIT

ONLY

£320 + VAT

(Single voice)

Plug in extra
Voices — Kit price £52 + VAT
(£48 + VAT if ordered with kit)

Cabinet size 31.1" x 19.6" x 7.6" rear 3.4" front

Kit also available as separate packs

ADSR IC		Pack		Price	Pack		Price
CEM 3310	£4.00	POLY 1	Pair of PCB's for multiplex cct. K.B.	£9.50	POLY 14	PCB for VOICE controls	£6.80
V60 10			contacts		POLY 15	Pots, switches, diodes, Cs for VOICE PCB	£4.80
VCO IC CEM 3340	£6.00	POLY 2	IC's, IC sockets, Rs. Cs. for multiplex cct.	€8.20	POLY 16	PCB for plug in voice	€8.20
CEIN SO 10		POLY 3	Superior quality keyboard	€32.25	POLY 17	Rs, Cs, presets, connectors for one voice	£16.30
0.1% 25 ppm		POLY 4	Contacts & bus bars	£12.00	POLY 18	ICs, IC skts, diodes for one voice	€27.50
M.F. Res	£0.50	POLY 5	Double sided plated through PCB for digital	£17.25	POLY 19	Transformer 0-120-240, 17-0-17, 0-7.7	£6.30
			control & pitch/gate generator cct		POLY 20	Pitch bend control	£3.90
0.5% 25 ppm		POLY 6	Rs, Cs, heat sink for fitting to Pack 5	£10.50	POLY 21	Misc parts eg. jack sockets, knobs, mains switch etc.	£13.00
M.F. Res	£0.25	POLY 7	IC's, IC sockets, diodes for firting to Pack 5	£31.30	0017100		
		POLY 8	Double sided mother board (for plug in voices)	£18.90	POLY 22	Ribbon cable, ribbon cable connectors, mains cable	£8.45
30 ppm multilaye		POLY 9	Rs, Cs, connectors for mother board	£14.10	0011100		
ceramic cap	£0.50	POLY 10	IC's IC sockets, Trs, heat sinks for mother	£13.10	POLY 23	Fully finished metalwork and fixing parts	€25.60
			board		POLY 24	Solid teak cabinet	£25.80
ICs and details		POLY 11	PCB for master controls (left of section marked VOICES)	£18.80	voice instru	or individually purchased packs for single iment	£355.15
packs in our FREE		POLY 12	ICs, IC sockets, diodes, Trs, Rs, Cs for master control PCB	£9.30	30 Complete kit for 4 voice expander including connectors		
CATALOG	UE	POLY 13	Pots, Switches for master control board	£11.80	All price	es VAT exclusive.	

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- Fill memory.
- Search memory
- Two save and three load routines
- Floppy disc vector.
- Break handling routine.
- Tabular display of memory

10 PRINT "FULL" 20 PRINT "SCREEN" 30 PRINT "TING"







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New! Sinclair ZX81 Personal Computer.

Kit: £49.95 complete

Reach advanced Reach advanced computer comprehension in a few absorbing hours

1980 saw a genuine breakthrough - the Sinclair ZX80, world's first complete personal computer for under £100. At £99.95, the ZX80 offered a specification unchallenged at the price.

Over 50,000 were sold, and the ZX80 won virtually universal praise from computer professionals

Now the Sinclair lead is increased: for just £69.95, the new Sinclair ZX81 offers even more advanced computer facilities at an even lower price. And the ZX81 kit means an even bigger, saving. At £49.95 it costs almost 40% less than the ZX80 kit!

Lower price: higher capability With the ZX81, it's just as simple to teach yourself computing, but the ZX81 packs even greater working

capability than the ZX80.

It uses the same micro-processor, but incorporates a new, more powerful 8KBASICROM-the 'trained intelligence' of the computer. This chip works in decimals, handles logs and trig, allows you to plot graphs, and builds up animated displays

And the ZX81 incorporates other operation refinements - the facility to load and save named programs on cassette, for example, or to select a program off a cassette through the keyboard.

Higher specification, lower pricehow's it done?

Quite simply, by design. The ZX80 reduced the chips in a working computer from 40 or so, to 21. The ZX81 reduces the 21 to 4!

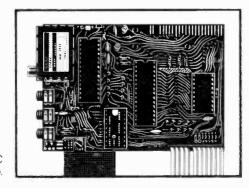
The secret lies in a totally new master chip. Designed by Sinclair and custom-built in Britain, this unique chip replaces 18 chips from the ZX80!

complete

Kit or builtit's up to you!

The picture shows dramatically how easy the ZX81 kit is to build: just four chips to assemble (plus, of course the other discrete components) - a few hours' work with a fine-tipped soldering iron. And you may already have a suitable mains adaptor - 600 mA at 9 V DC nominal unregulated (supplied with built version).

Kit and built versions come complete with all leads to connect to your TV (colour or black and white) and cassette recorder.

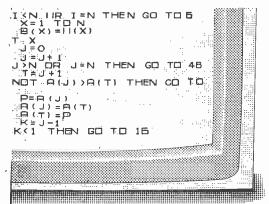


New Sinclair teach-yourself BASIC manual

Every ZX81 comes with a comprehensive, speciallywritten manual -a complete course in BASIC program-

ming, from first principles to complex programs. You need no prior knowledge - children from 12 upwards soon become familiar with computer operation.

Proven micro-processor, new 8K BASIC ROM, RAM - and unique new master chip.



New, improved specification

 Z80A micro-processor – new faster version of the famous Z80 chip, widely recognised as the best ever made.

- Unique 'one-touch'
 key word entry:
 the ZX81
 eliminates a great
 deal of tiresome
 typing. Key words
 (RUN, LIST, PRINT,
 etc.) have their own
 single-key entry.
- Unique syntaxcheck and report codes identify programming errors immediately.
- Full range of mathematical and scientific functions accurate to eight decimal places.
- Graph-drawing and animateddisplay facilities.
- Multi-dimensional string and numerical arrays.
- Up to 26 FOR/NEXT loops.
- Randomise function useful for games as well as serious applications.
- Cassette LOAD and SAVE with named programs.
- 1K-byte RAM expandable to 16K bytes with Sinclair RAM pack.
- ◆ Able to drive the new Sinclair printer (not available yet – but coming soon!)
- Advanced 4-chip design: microprocessor, ROM, RAM, plus master chip – unique, custom-built chip replacing 18 ZX80 chips.

SINCIBIN ZX81

Sinclair Research Ltd,

6 Kings Parade, Cambridge, Cambs., CB2 1SN. Tel: 0276 66104. Reg. no: 214 4630 00

ETI JUNE 1981

lf you own a Sinclair ZX80...

The new 8K BASIC ROM used in the Sinclair ZX81 is available to ZX80 owners as a drop-in replacement chip. (Complete with new keyboard template and operating manual.)

With the exception of animated graphics, all the advanced features of the ZX81 are now available on your ZX80 – including the ability to drive the Sinclair ZX Printer.

Coming soonthe ZX Printer.

Designed exclusively for use with the ZX81 (and ZX80 with 8K BASIC ROM); the printer offers full alphanumerics across 32 columns, and highly sophisticated graphics. Special features include COPY, which prints out exactly what is on the whole TV screen without the need for further instructions. The ZX Printer will be available in Summer 1981, at around £50 – watch this space!



16K-BYTE RAM pack for massive add-on memory.

Designed as a complete module to fit your Sinclair ZX80 or ZX81, the RAM pack simply plugs into the existing expansion port at the rear of the computer to multiply your data/program storage by 16!

Use it for long and complex programs or as a personal database. Yet it costs as little as half the price of competitive additional memory.



How to order your ZX81

BY PHONE – Access or Barclaycard holders can call 01-200 0200 for personal attention 24 hours a day, every day. BY FREEPOST – use the no-stampneeded coupon below. You can pay by cheque, postal order, Access or Barclaycard.

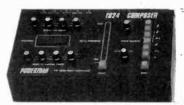
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TRANSCENDENT 2000

COMPLETE KIT ONLY £168.50 + VAT!

Designed by consultant Tim Orr (formerly synthesizer designer for EMS Ltd) and featured as a constructional article in ETI, this live performance synthesizer is a 3 octave instrument transposable 2 octaves up or down giving sweep control, a noise generator and an ADSR envelope shaper. There is also a slow oscillator, a new pitch detector, ADSR repeat, sample and hold, and special circuitry with precision components to ensure tuning stability amongst its many features.



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SINGLE BOARD SYNTHESIZER



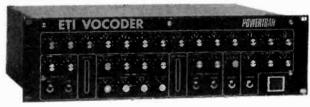
Cabinet size 24,6" x 15,7" x 4,8" (rear), 3,4" (Front).

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KIT INCLUDES FREE FOOT CONTROL AND TEST OSCILLATOR



Panel size 19.0" x 5.25". Depth 12.2".

Featured as a construction article in **Electronics Today International** this design enables a vocader of great versatility and high intelligibility to be built for an amazingly low price.

14 channels are used to achieve its high intelligibility, each channel having its own level control. There are two input amplifiers, one for speech either from microphone or a high level source e.g. mixer or cassette deck and one for external excitation (the substitution signal) from either high or low level sources. Each amplifier has its own lever control and a rather special type of tone control giving varying degrees of bass boost with treble cut or treble boost with bass cut. The level of the speech and excitation signals are monitored by LED PPM meters with 10 lights — 7 green and 3 red which indicate the level at 3dB steps. There are three internal sources of excitation — a noise generator and two pulse generators of variable frequency and pulse width. Any of the internal sources and the external source can be mixed together. There is a voiced / unvoiced detector which substitutes noise for the excitation signal at the points in speech where the vocal chord derived sounds of the speaker are substituted for by the unvoiced sounds of sibilants, etc. There is a slew rate control which smooths out the changes in spectral balance and amplitude enabling a change of the speech into singing or chanting and other special effects. A foot switch is provided to permit a complete freeze in spectral balance and amplitude whenever required. An LED on this indicates when the freeze is in operation.

the vocal chord derived sounds of the speaker are substituted for by the unvoiced sounds of sibilants, etc. I here is a slew rate control which smooths out the changes in spectral balance and amplitude enabling a change of the speach into singing or chanting and other special effects. A foot switch is provided to permit a complete freeze in spectral balance and amplitude whenever required. An LED on this indicates when the freeze is in operation.

An output mixer allows mixing of the speech, external excitation and vocoder output. The majority of the components fit into the large analysis/synthesis board with the rest on 8 much smaller boards with the controls and sockets mounted on them for ease of construction. Connectors are used for the small amount of wiring between the boards. The kit includes fully finished metalwork, professional quality components (all resistors 2% metal oxide) nuts, bolts, etc. — even a 13A plugl

TRANSCENDENT DPX

MULTI-VOICE SYNTHESIZER

Another superb design by synthesizer expert Tim Orr published in Electronics Today International

COMPLETE KIT ONLY £299 + VAT!



Cabinet size 36.3" × 15.0" × 5.0" (rear) 3.3" (front)

The Transcendent DPX is a really versatile 5 octave keyboard instrument. These are two audio outputs which can be used simultaneously. On the first there is a beautiful harpsichord or reed sound—fully polyphonic, i.e., you can play chords with as many notes as you like. On the second output there is a wide range of different voices, still fully polyphonic. It can be a straightforward piano as a honky tonk piano or even a mixture of the twol Alternatively you can play strings over the whole range of the keyboard or should you prefer — strings on the top of the keyboard and brass as the lower end (the keyboard is electronically split after the first two octaves) or vice-versa or even a combination of strings and brass sounds is simultaneously. And on all voices you can switch in circuitry to make the keyboard touch sensitive! The harder you press down a key the louder it sounds—just like an accoustic piano. The digitally controlled multiplexed system makes practical touch sensitivity with the complex dynamics law necessary for a high degree of realism. There is a master volume and tone control, a separate control for the brass sounds and also a vibrato circuit with variable depth control together with a variable delay control so that the vibrator groups in only after waiting a short time after the note is struck for even more realistic string sounds.

vibrator comes in only after waiting a short time after the note is struck for even more realistic string sounds.

To add interest to the sounds and make them more natural there is a chorus / énsemble unit which is a complex phasing system using CCD (charge coupled device) analogue delay lines. The overall effect of this is similar to that of several acoustic instruments playing the same piece of music. The ensemble circuitry can be switched in with either strong or mild effects.

As the system is based on digital circuitry digital data can be easily taken to and from a computer (for storing and playing back accompaniments with or without pitch or key change, computer composing, etc., etc.).

Although the DPX is an advanced design using a very large amount of circuitry, much of it very sophisticated, the kit is mechanically extremely simple with excellent access to all the circuit boards which interconnect with multiway connectors, just four of which are removed to separate the keyboard circuitry and the panel circuitry from the main circuitry in the cabinet. The kit includes fully finished metalwork, solid leak cabinet, professional quality components (all resistors 2% metal oxide), nuts, bolts, etc., even a 13A plugl

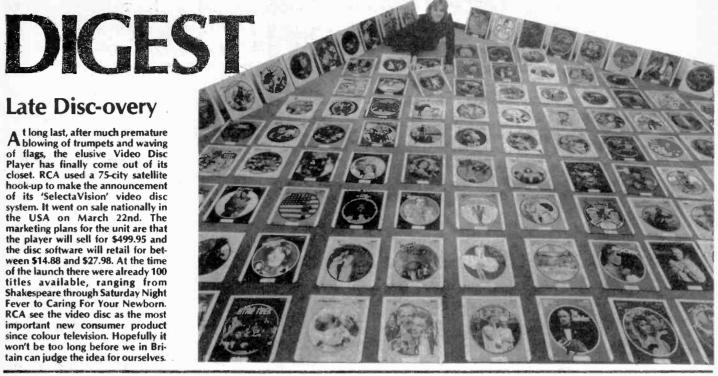
POWERTRAN

MANY MORE KITS ON PAGE 100. MORE KITS AND ORDERING INFORMATION ON INSIDE FRONT COVER

All projects on this page can be purchased as separate packs, e.g. PCBs, components sets, hardware sets, etc. See our free catalogue for full details and prices

Late Disc-overy

A t long last, after much premature blowing of trumpets and waving of flags, the elusive Video Disc Player has finally come out of its closet. RCA used a 75-city satellite hook-up to make the announcement of its 'SelectaVision' video disc system. It went on sale nationally in the USA on March 22nd. The marketing plans for the unit are that the player will sell for \$499.95 and the disc software will retail for between \$14.88 and \$27.98. At the time of the launch there were already 100 titles available, ranging from Shakespeare through Saturday Night Fever to Caring For Your Newborn. RCA see the video disc as the most important new consumer product since colour television. Hopefully it won't be too long before we in Britain can judge the idea for ourselves.



Computing In Sinc

Uncle Clive has very definitely done it again with his new ZX 81 Personal Computer. It has many improved features over the ZX 80. For instance, it is based on a four-chip design and is constructed in durable black ABS plastic, making it far more sturdy than its flimsy white pre-decessor. The masterchip is manufactured by Ferranti and is custom-built to replace 18 chips in the ZX 80. The 81 has 8K BASIC ROM and incorporates all the features of its forerunner. The advanced features are a print drive facility and the ability to operate in two software selectable modes, 'fast' and 'normal' (fast being four times normal speed). In 'normal', the ZX 81 will compute and

Ende le le

ZX81

EI

display simul-

a compare to the comp

taneously giving continuous flickerfree graphics - a major deficiency on the ZX 80. The new 40-key touchsensitive membrane keyboard gives the equivalent of 91 keys. graphics mode enables an additional 20 graphical and 54 inverse video characters to be entered directly from the keyboard. Programs can be loaded and saved using a conventional recorder. The cassette interface facility has been improved and programs are given names so that you may use the computer to search through a tape for the required program. A comprehensive 200-page tome is provided which includes a new course in BASIC programming. The ZX 81 is available readyassembled for £69.95, in kit form for £49.95 but with separate mains adaptor for a further £8.95. There are also some new accessories from Sinclair. The new 8K BASIC ROM for the 81 will soon be available to ZX 80 owners as a drop-in replacement chip, with a new keyboard template operating manual. Price is £19.95. There is also a 16K byte RAM pack for both the 80 and the 81 for £49.95, which can be used for program storage or as a database. There will also be a printer launched later this year for around £50, its special feature being a facility called 'copy' which prints out exactly what is on the TV screen. There are also four cassettes of software

The add-on 16K RAM pack for the ZX 81.

immediately available

at £3.95 each. The

ZX 81 will be sold

by mail order only.

Surging Ahead Suhner Electronics have announced a range of EMP protectors designed to limit surge voltages in coaxial systems. Their effectiveness is mainly due to their rapid response time of 3 ns, as opposed to a typical delay of 10 ns in other protectors. After a dangerous surge voltage such as that found in a lightning strike, the rare gas-filled protector sparks over, shorting the inner and outer conductor and allowing the surge current to flow to ground. The equipment is back in operation immediately afterwards, with full protection maintained. The series 3401 EMP protectors have the form of a coaxial feedthrough adaptor for wall-mounting.

A very low and stable grounding contact between wall-mounting and protector body is achieved by a V-groove washer made from soft copper. The protectors are accurately matched to 50 R line impedance and the shunt capacitance of the UC protector is carefully compensated to ensure minimum reflections when operating at the commonly used communication bands. The range of protectors is specially designed and tested for military applications, but has gained a broad-band acceptance as excellent lightning protectors for civil communications and antenna equipment. The protectors can be obtained from Suhner Electronics Ltd, Telford Road, Bicester, Oxon OX7 OLA.



Measuring Up Data Precision have launched a

new 41/2 digit multimeter called the DP255 with LCD display and a basic accuracy of .03%. It is a highly accurate, average sensing, calibrated RMS multimeter with a rechargeable Ni-Cad battery pack, which will operate for up to 100 hours between recharging. The meter measures AC and DC volts and

current as well as resistance. Voltage resolution is 10uV, current resolution 10nA and resistance resolution 10.1R. All 25 ranges are selected with two front panel rotary switches, one for function and the other for range. The unit weighs less than 1.3lb, has 0.4" high digits and accepts banana plug connection. For more information contact Farnell International Instruments Ltd, Sandbeck Way, Wetherby, West Yorks LS22 4DH.

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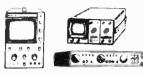
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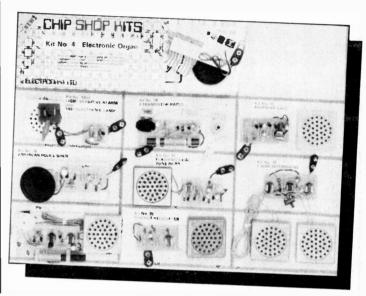


Computer Games

f you, like many people, are a computer games addict, you'll be interested to hear about two products which Hanimex are launching this year. The first of these is the Interton VC4000 Microprocessor Video Computer which consists of a main video computer console and an initial software back-up of 22 varying cassettes (with a promise of 10 more by June 81). Included in the software list are favourites like 'Space Invaders' and 'Space Wars', as well as new games like 'Cockpit', an aircraft simulation game. The VC4000 is expected to retail at about £99.95 and the cassettes at £16.95 in the basic range. Certain of these contain up to 60 games each. For the more microcomputerminded among us, Hanimex will be launching a combined Microprocessor TV Computer and Microcomputer which will retail for under £200, to become available in April or May. It has a small keyboard and fully programmable RAM incorporated in the console, together with four resident TV games plus a range of cartridges and software. Hanimex UK is at Hanimex House, Dorcan, Swindon SN3 5HW, Wilts.

Cheap Kits

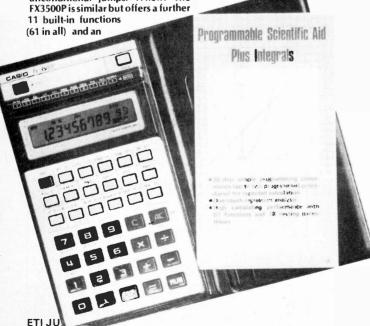
or the first time in Britain a new range of practical electronic kits retailing for under £5 will be available from high street hobby shops. They are called Chip Shop Kits and there are over 20 different versions to choose from. Included in the range are projects as diverse as a soldering iron, electronic organ, four-way transistor radio and a lie detector! Each kit contains all the components you need, and all you have to do is wield your soldering iron and add a 9 V PP3 battery. Stepby-step instructions are included with detailed educational notes about the individual circuit and advice on soldering techniques. Keep your eyes open for them in your local hobby shop, or contact Electroni-Kit Ltd, Rectory Court, Chalvington, Hailsham, East Sussex BN27 3TD.



Totting Up

C asio are, once again, introducing a new series of three LCD calculators. The FX2700P is a slimline (7 mm thick) pocket machine with 50 built-in functions: polar/rectangular, random number, 18 parenthesis (nestable up to six levels) fixing of significant digits and of decimal places, standard deviation, etc. Capacity is eight digits, plus two-digit exponent. There are seven memories plus programming to a limit of 38 steps, complete with conditional and unconditional jumps. Whew! The FX3500P is similar but offers a further

extra two-digit capacity. The FX180P is closely related, featuring 55 functions (no hyperbolics) and the same standard deviation, linear regression, integration and 38-step two program facilities. It is 20 mm thick, to accommodate AA size batteries which give 8,000 hours' continuous operation. The FX2700P and FX180P both have recommended retail prices of £22.95 and the FX3500P is £25.95, but, as usual, if you shop around you can probably obtain them for less.



Music From A/D

MI has made available a number of albums recorded using digital techniques. These include the works of Mahler, Strauss, Debussy, Bach, Wagner and Scott Joplin. Further information is obtainable from EMI Ltd, 20 Manchester Square, London W1A 1ES.

Booking Up
We've just received another four
paperbacks from the prolific
publisher Mr Babani. The first of these is the 'International Transistor Equivalents Guide' by Adrian Michaels. This book lists over 22,000 transistors produced by more than 100 manufacturers, and their equivalents (but then you'd guessed that from the title, hadn't you?). No characteristics are given but each entry has information on transistor type, polarity, manufacturer(s), the European, American and Japanese equivalents and typical applications. The guide is priced at £2.95.

VMOS Projects', by R.A. Penfold, contains 96 pages of circuits using VMOS power FETs. The book opens with a discussion of the characteristics of VMOS devices, comparing them to other types of transistors, and goes on to describe audio circuits, sound generators and alarms, DC control circuits and signal control circuits — in all, a total of 33 projects. Each circuit suggestion is accompanied by a thorough explanation and a parts list. The book concludes with a table giving the pinouts of all the semiconductors used. VMOS Projects costs £1.95.
The third tome is 'Digital IC Pro-

jects', by F.G. Rayer (91 pages); this book is divided into three sections.

The first gives a general guide to components and construction, including IC orientation and pinouts for Nixie tubes and seven-segment displays. The second section deals with power supplies for the projects, explaining the principles of voltage regulation and giving suitable cir-cuits. Following this are the projects themselves, ranging from roulette and digital dice to counters, stopclocks and test gear. This book also costs £1.95.

Finally we have 'Electronic Synthesiser Projects' by M.K. Berry. In its 81 pages it covers most aspects of electronic music and gives simple circuits for the experimenter to build. The first of these is a singlechip synthesiser based on the SN76477N sound generator IC, and later chapters describe an analogue delay line, a programmable analogue sequencer, two VCOs, an envelope shaper and a power supply. The final chapter explains how the various circuits can be used together. Each project has a parts list, a suggested PCB and a component overlay. All this can be yours for only £1.75. All these books should be readily available in bookshops, but in case of difficulty you can contact Bernard Babani (Publishing) Ltd at The Grampians, Shepherds Bush Road, London W6.

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Stacks Of Time

At the Home Video Show in March, those awfully nice Sony people launched a new add-on accessory for their C7 video recorder. With the new generation of programmable video recorders owners have found themselves with a problem — the tapes simply aren't long enough! So Sony have come up with the Betastack which gives the owner up to 13 hours of unattended recording — a great help if you are recording — a great help if you are recor-

The Betastack can be fitted to the C7 by any Sony dealer for under £150, or a new purchaser can buy it already fitted for an extra £129. For further information, contact your Sony dealer. Sony were also selling demonstration tapes at their stand showing how best to make your own video films. The tapes comprise of 30 minutes of demonstration with the rest of the tape left blank for you to record your own efforts. The tapes will still be available after the show for £6.50 so if you want to get hold of one contact the Sony Tape Group, Pyrene House, Sunbury on Thames, Middly

RTVC Amp

n last month's Kit Review we looked at the RTVC 10 + 10 Stereo Amplifier and mentioned some errors in the instructions. As promised we are publishing their reply, which appears below.

Dear Editor.

First let us thank you for the review of our 10 + 10 Stereo Amplifier Kit, we much appreciate your comments on "good value for money". Thank you also for pointing out the errors in the instructions. On receipt of your copy we immediately printed a correction slip which has been included with every kit sold since.

As we are sure you are aware such errors can get through the checking procedures — we notice that your own projects are often followed by corrections and we are surprised to find you are also guilty of failing to provide adequate or correct information on mains connections. For instance in your Soldering Iron project in the May issue no mains wiring information is given, other than the circuit diagram.

It is very easy to criticise such mistakes, which can so easily occur; however, we have now been able to correct ours (thanks to your review) and we believe we can thus maintain our tradition of supplying good quality kits which represent very good value for money.

As you have said the RTVC 10+10 "must be the cheapest stereo amplifier you can buy".

Yours faithfully, L. Cohen P.S. I trust that in view of your final paragraph under In Use (p.32 May issue) you will print the above reply in full.



Night Vision

ith the number of burglaries on With the number of burgantees the increase, why not take advantage of the Night Owl Radar Controlled Courtesy Light and Intruder Detector from Loadpoint? The Night Owl can protect an egg-shaped area up to a forward range of 100 feet using radar. When the system detects a moving object — a person or a vehi-cle, for example — the internal lights are automatically turned on and will stay on until the last movement has ceased. The unit is programmed to accept only continuous movement

objects like doors, curtains and
tree branches will not trigger the device. The detection range is adjustable from 6 to 100 feet and the light duration can be set from 1/2 to 10 minutes. If an audible alarm or extra light is required an extra set of contacts are already built into the unit. The device can be easily fitted to a 220/240 V mains supply. The unit costs £150, and further details can be obtained from Loadpoint Ltd, Chelworth Industrial Estate, Cricklade, Swindon, Wilts SN6 6HE.

Protect and Survive

This year's Ideal Home Exhibition did not simply cover the luxury aspects of life in the 1980s. Protect and Survive Monthly magazine staged its first display of nuclear war survival equipment this year. The magazine shows some of the latest equipment in the survival business, including NBC suits designed to protect the wearer against radioactive fallout dust after a nuclear explo-

sion. Also included were survival foods, equipment and material for use in a fallout bunker, the latest in radiation meters and a display of fallout shelter ventilation equipment from Switzerland. The magazine's slogan is 'You can survive a nuclear war'. The publisher believes that Protect and Survive covers subjects which "until recently, have not been brought to the public's attention". For more information on this interesting subject contact Protect and Survive Monthly, 80 Fleet Street, London EC4Y 1EL.

Double-sided

Sharp Electronics (UK) Ltd has developed a new music centre which can play both sides of a record without turning it over! The VZ3000 is a major breakthrough in convenience for music lovers: you can even repeat the process for non-stop music. The system comprises a fully

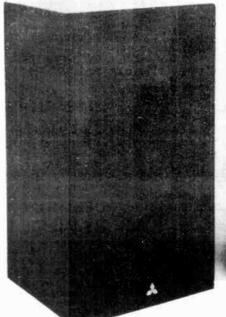
automatic 2-speed belt-driven vertical 12" turntable that detects the speed and size of the record, and its two linear tracking arms with VM cartridges allow both sides to be played automatically. The unit also incorporates a LW/MW/FM stereo radio with a four-track, two-channel stereo cassette deck and speakers. It is expected to retail at £325.

Computing With Aunty

The BBC are currently planning a series of computer literacy programmes for broadcast next lanuary. Aunty Beeb will also be launching its own microcomputer in conjunction with the series which will give the viewer the opportunity of gaining practical experience of computing in the home. The micro is a condensed version of the forthcoming Proton from Acorn Computers Ltd. It will be built to the BBC's very high specification and will be capable of being linked to teletext transmissions by using an add-on receiver. It will also have a sophisticated graphics display which will allow you to play your favourite games and also develop your own pictures. Many other features are included in the unit, which is expected to retail for around £200 with the teletext adaptor costing a further £100. More information on this exciting project will appear on these pages as soon as it is available.

Speaking Out
A sa follow-up to the Mitsubishi DS

As a follow-up to the minimum and a safety for the speaker reviewed in last month's Audiophile, comes the DS 25 Mark II. It is a two-way bass reflex system handling 100 W. The 10" woofers are almost identical to those in the D\$ 32 and the port on the baffle panel has been specially treated to ensure additional damping of residual resonant vibrations to give a clean bass response. It is set in heat-resistant plastic to permit high input levels without insulation breakdown. The voice coil features anti-resonance apertures to reduce unwanted resonance and diste The resonant frequency tweeters has been set well below the crossover frequency to ensure a speaker response free of uneven dispersion and 'beaming'. The centre cap is a titanium dome with a parabolic profile and a four-step at-tenuator is featured to enable the user to adjust tweeter response to suit the particular listening environment. The retail price of this pair of speakers is £153. For further details contact Mitsubishi Electric (UK) Ltd, Otterspool Way, Watford, Herts WD2 8LD.





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THE RADOFIN TELETEXT ADD-ON ADAPTOR
Plug the adaptor into the aerial socket of your colour TV and receive the CEEFAX and ORACLE television information services.

THIS NEW MODEL INCORPORATES:

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Fast add listing PRINTER/
CALCULATOR: 2 lines per second. 10 digit capacity. Uses normal adding machine rolls. Battery or mains operated.
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Plays 24 different tunes with separate speed control and volume control. Select the most appropriate tune for your visitor, with appropriate tunes of the year!

MATTEL T.V. GAME



The most advanced T.V. game in the world 20 cartridges available Add. of KEYBOARD coming 199, sinc. soon to convert the community of the Add. MATTEL to a home computer with 16K RAM, fully expandable and programmable in Microsoft Basic. Other accessories will be available later in the year.

THE OLYMPIA **POST OFFICE APPROVED**



TELEPHONE ANSWERING MACHINE

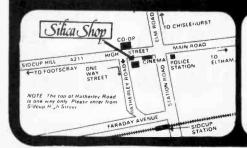
WITH REMOTE CALL-IN BLEEPER

This telephone answering machine is manufatured by Olympia Business. Machines, one of the largest Office Equipment manufacturers in the U.K. It is fully POST OFFICE APPROVED and will answer and record messages for 24 hours a day. With your remote call-in bleeper you are in the world. The remote call-in bleeper activates the Answer/Record Unit, which will at your command repeat messages, keep or erase them, and is activated from anywhere in the world, or on your return to your home or office. The machine can also be used for message referral, if you have an urgent appointment, but are expecting an important call, simply record the phone number and location where you can be reached. With optional extra bleepers (£ 13 each) this facility can be extended to colleagues and members of the family Using a C90 standard cassette you can record as many as 45 messages. The announcement can be up to 16 seconds long and the incoming message up to 30 seconds long.

The machine is easy to install and comes with full instructions, it is easily wired to your junction box with the spade connectors provided or alternatively a jack plug can be provided to plug into a jack socket. Most important, of course, is the fact that it is fully POST OFFICE APPROVED.

The price of £135 (inc. V.A.f.) includes the machine, an extra-light remote call-in Bleeper, the microphone message tape, A/C mains adaptor. The unit is 9%%%2%2% and is fully guaranteed for 12 months. The telephone can be placed directly on the unit — no additional desk space is required.

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rates of interest
PART EXCHANGE SCHEME pscalable on 2nd hand machine chemic charge control of the control of th SILICA SHOP LIMITED Dept. ET16 1-4 The Mews, Hatherley Road, Sidcup, Kent DA14 4DX Telephone: 01-301 1111 or 01-309 1111





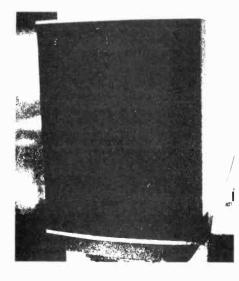
Here's one for all you DJs who seek technical perfection in their equipment. The DJ90 Stereo Mixer has two stereo inputs for magnetic cartridges, an AUX input (also stereo) with a flat response, and a microphone input. There are slider controls for mixing the three music inputs together and it's possible to pan automatically between the record inputs at either fast or slow rates. Processing of the music signal is available in the form of a five-section graphic equaliser with two-octave spacing, plus a special 'beat lift' device. A voice-over unit is provided, as well as an override button for interrupt announcements. A stereo headphone socket on the monitor section enables pre-fade listening to any of the music inputs, while an LED PPM displays the signal level.

SYSTEM A AMPLIFIER

It's classy and it's Class A. The ETI System A has been designed so that the home constructor can build a hi-fi system offering a comparable performance to any ready-built unit — and with metalwork kits available, it will look just as good too. The preamp offers ultralinear operation, choice of input modules, minute output impedance and truly amazing sound quality. The power amps are capable of delivering 150 W (RMS!) into an 8R load, with distortion so low you have to hear it to believe it. System A is simply the best.

BATTERY CHARGER

Not another one, did we hear you say? Quietly, please, or you'll hurt its feelings — after all, this is a *smart* charger. It knows when you've connected dud batteries or ones with wrong voltage, and ignores them. If it's satisfied with the batteries, it can pump up to 4 A into them until they're charged, then switch to trickle-charge to keep them topped up. There are LEDs to indicate 'on', 'charging' and fault conditions. Smart, eh?



This virtually indistinguishable blob will revolutionise the hi-fi field. This is the new Quad loudspeaker — rumoured since the sixties.

Next month ETI presents an exclusive in-depth report on this new transducer. You can only read the details here. Others will give you supposition — ETI gives you the FACTS!

Don't miss this exclusive article!

ELECTRONICS IN SUBMARINES

Submarine warfare has been a part of military strategy since before the First World War. Since then the effectiveness of these fighting vessels has been dependent on advances in electronics, such as sonar, radar and computer control of tracking and missile guidance. This feature examines the role of the submarine in naval warfare and the important implications for the British nation.

ZX81

Last month saw the release of Clive Sinclair's latest personal computer, the ZX81. Microbasics will be taking a look at this machine and discussing how its capabilities suit the beginner. If you're thinking of making a start in computing, don't miss the July ETI.



LOOK OUT FOR THE JULY ISSUE ON SALE JUNE 5th

REMOTE CONTROL COMPONENTS AND KITS



LD271 IR Emitting diode	.36
SFH205 Photodiode Detector	.90
SL480 IC Pulse Amp.	1.70
SL490 32 Command Encoder/	
transmitter	2.40
ML922 10-channel receiver	
+ 3 analogue outputs	4.20



1.40
1.40
1.40
.20

ML925. A decoder designed for model/toy control, providing a 2-speed drive motor and three position latched steering system or a vehicle with momentary action steering and a third motor, eg. gun turret, winch, etc. Outputs also available for other facilities such as horn, turn indicators, headlights, etc. 2.10

winch, etc. Outputs also available for other ractifiles such as norn, furn indicators, necessity is excessed in the test of the such as norn, furn indicators, necessity is excessed in the set of the such as a such as

MK11 — 10 On-OH Channel IR RECEIVER with 3 analogue outputs (0-10V) for controlling such functions as lamp brightness, volume, tone etc. Other functions include an on-/standby output and a toggle output, which may be used for sound muting. Based on ML922 decoder IC Includes its own mains supply 12,00 MK12 — 16 Channel IR RECEIVER. For use with the MK8 kit with 16 on-/off outputs which with further interface circuitry, such as relays or triacs, will switch up to 16 items of equipment on or off remotely. Outputs may be latched or momentary, depending on whether the ML926 or ML928 is specified. Includes its own mains supply. 11.95

NEW

MK13 — 11-Way KEYBOARD. For use with MK8 and MK11 kits.

Transmits programme step + and -, analogue + and - (3), mute.

normalise analogue outputs, and on/Standby

4.35

0		M
926 16-channel receiver 4		м
momentary binary outputs	1.40	-
928 16-channel receiver, 4		
latched binary outputs	1.40	- 82
929 16-channel receiver, 4		m
latched binary outputs	1.40	ш
o-on Plastic Reflector for		-
IR LEDs, increases range	.20	

11:12:

Our new TDR300K Touch Dimmer Kit will ensure that you are. Based on our highly successful TD300K touch controlled dimmer kit, the TDR300K incorporates an infra red receiver, enabling the lamp brightness to be varied and switched on or off by touch or remotely by means of a small hand held transmitter. The complete kit, which includes easy to follow instructions, will fit into a plaster depih box and the plastic front plate has no metal pads to touch, ensuring complete safety. Even a neon is included to help you locate the switch in the dark.



ARE YOU SITTING COMFORTABLY?

In years to come everyone will be selling remote control dim-mers, but you can have your TDR300K kit now for ONLY £14.30 for the dimmer unit and £4.20 for the transmitter.

For the more athletic of you, the TD300K Touchdimmer kit is still available at £6.50 and the TDE/K Extension kit, for 2-way switching etc., is £2. DON'T FORGET to add 50p P&P and 15% VAT to your total purchases. purchase

24HR. CLOCK/APP. TIMER KIT



Switches any appliance up to 1KW on and off at preset times once per day Kit contains AY.5-1230 IC. 0.5° LED display mains supply, display trivers switches; LEDs triac PCBs & full instructions.

CT1000K Basic Kit

CT1000KB with white box (56/131 - 71) £17.40

DIGITAL VOLTMETER THERMOMETER KIT



Based on the ICL 7106 This kit contains a PCB resistors, presets, capacitors diodes IC and 0.5" Injuried risk and state of the state of

£19.50

INTEGRATED CIRCUITS

555 Timer 741 Op Amp AV-5-1230/2 Clock AY-5-1230/2 Clock / Timer AY-3-1270 Thermometer CA3080 Transconductance Op Amp CA3130 CMOS Op Amp ICL7106 DW (LCD drive) LM377 Dual 2 W Amp LM380 2W Audio Amp LM380 2W Audio Amp LM380 2W Audio Amp LM380 550mW low voltage Amp LM380 550mW low voltage Amp LM380 Follow Fore Freamp LM381 Store Laster (14 pin) LM390 LED Flasher / Socillator LM391-1 Thermometer LM3915 Dot: Bar Driver (log) MM74C911 4 digit display controller MM74C926 4 digit counter with 7 seq or S5668 Touchdimmer S1440 AC power control SN76477 Complex Sound Generator T8A800 SW Audio Amp TRA810AS 7W Audio Amp TRA810AS 7W Audio Amp TRA810AS 7W Audio Amp TRA91 JFET Op Amp TL081 JFET Op Amp TL081 JFET Op Amp TL034 LTIMET TL034 L	21p 119p 62.80 64.50 68.20 75p 67.00 61.45 63.50 61.00 75p 61.00 60p 61.20 62.10 62.20 62.50 62.50 62.50 62.50 62.50 62.85 62.50 62.85 62.85 62.85 62.85 62.85 62.85 62.85 62.85 63.
All ICs supplied with Data Sheets Data Sheets only — per device	100

CMOS							
1000	17p	4019	42p	4069	19p		
1001	18p	4023	22p	4070	19p		
1002	18p	4025	21p	4071	18p		
4007	17p	4026	£1,30	4077	26p		
4011	19p	4027	40p	4081	20p		
4012	170	4028	50m	4093	54p		
4013	38p	4040	80p	4501	240		
4015	75p	4049	38p	4511	90p		
4016	35p	4050	40p	4514	£1.80		
4017	700	4060	61.00	4516	C1.00		

CAPACITORS

Polyester 250V 0.01 0.002 0.033 0.047 0.068 0.1 0.15			6p 6p 7p 7p 7p 7p 7p 12p	0.2 0.3 0.4 0.6 1.0 1.5 2.2	3 7 8	12p 12p 15p 18p 24p 27p 31p
63v 25v	1 0 2.2 4.7 10 47 22 47 100 110 470	FRRRRAAAAA	5p 5p 6p 7p 12p 8p 10p 14p 15p 24p 34p	16V	100	R 5pp R 5pp R 5pp R 6p R 11pp R 14pp R 20pp A 8pp A 8pp A 12p
25V	um (beac 0.1 0.22 0.47 1.0 2.2 4.7	1)	9p 9p 9p 9p 10p 14p 20p	10V 6.3V 3V	22 33 47 100	20p 20p 27p 27p
100pF 220pF 470pF	ic 50V	4	lp lp	2	.000pF 200pF .700pF .000pF	4ր 4ր 5ր 5ր



EVERY DOOR SHOULD HAVE ONE

Whatever kind of door you have, our New Electronic Combination Lock will enable you to open it easily but make things very difficult for unwelcome visitors. The unit, which comes complete with a 10-way keypad, requires an easily remembered four digit code to be entered before the door can be opened, while the intruder has over 5,000 combinations to choose from. The code can be easily changed by means of a pre-wired plug and a momentary or latched output version can be made. The kit has even more uses in a car where it may be used to disable the ignition. Another useful feature is the Save Button. This stores the combination number, enabling the car to be used by authorised persons such as garage personnel without disclosing the code.



The complete kit measures 7 x 6 x 3 cms, deep and consumes a mere 40uA when not in use, and will drive a 5V to 15V (750mA) solenoid or relay coil (not supplied) directly. So why not treat your door to a new lock for

ONLY £10.50 + VAT

and think about all the keys you can lose or forget without ever locking yourself out.
As featured in PE, May '81.



LEDS

TRIACS 400V Plastic Case (Texas) 3A TIC206D 8A TIC226D 12A TIC236D 16A TIC246D 25A TIC263D 6A with trigger Q4006LT 8A isolated tab TXAL226B

VOLTAGE RECHLATORS

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6	Available in 5V, 12V versions.	& 15V
	78L series 100mA pos. 79L series 100mA neg. 78 series 1A pos	26p 60p 52p
LM 317T adjusta	bie 1.2V-37V 1.5A	£1.80
79 Series 1A ne	g-5 vol	75p

RECTIFIERS

THE OTHER DELIC	
1N4001 1A/50V	4р
1N4004 1A/400V	6p
1N5401 3A/50V	8p
1N5404 3A / 400V	10p
1A 50V Bridge	19p
1A 400V Bridge	220
3A 50V Bridge	48p
3A 400V Bridge	58p
(Bridge rectifiers are miniature PCB mou	unting types)

VMOS POWER FETS

VN10KM 0 5A / 60V (TO92)	52p
VN66AF 2A : 60V (TO220)	88p
TRANSISTORS	

BC108	8p	BF (50)	20p
BC182	8p	TIP31A	40p
BC182L	8p	TIP32A	40p
BC212	9p	2N3055	45p
BC212L	9 _D	2N3442	90p
BC337	16p	2N3819	20p

MINI TRANSFORMERS



Standard mai	
primaries 240	Va.c.
100mA secon	daries
6-0 6V	80p
9-0-9V	85p
12-0-12V	90p
Opto isolated	
TRIAC	
M O C 3	0 2 0
0.6A/400V	£1.10

9p 12p 12p 9p 12p 12p 3p 0.2" clips 3p Rectangular Red 16p Rectangular Green 17p Rectangular Yellow 17p Flat Face Rectangular. Triangular. Arrowhead Square

DISPLAYS



MP463 0.5" c.c. Red, four digit multiplexed clock display £2.20

DILLIC SOCKETS

8	pin	8p	18 pin	17p
	pin	12p	28 pin	24p
16	pin	14p	40 pin	36p
So	Idercon Pir	is		50p/100

DISCO LIGHTING KITS

Each unit has 4 channels (rated at 1KW at 240V per channel) which switch lamps to provide sequencing effects, controlled manually or by an optional opto solated audio input.

OL 1000K

DL 1000K
This kit features a bi-directional sequence, speed of sequence and frequency of direction change being variable by means of potentiometers. Incorporates master dimming control. £14.60 DL2100W.
A lower cost version of the above, featuring undirectional channel sequence with speed variable by means of a preser pot Outputs switched only at mains zero crossing points to reduce radio interference to minimum. £8.00 Optional Opto Input

Optional Opto Input DLA1

60p

MINI KITS

These kits form useful subsystems which may be incorporated into larger designs or used alone. Kits include PCB, short instructions and all com-

ponents
MK1 TEMPERATURE CONTROLLER /
THERMOSTAT
Uses LM3911 IC to sense temperature (80°C
max) and triac to switch heater
1KW £4.

MK2 SOLID STATE RELAY

MK2 SOLID STATE NELAT leaf for winching motors, lights, heaters, etc from logic. Opto-isolated with zero voltage switching. Supplied without triac Select the required triac from our range. £2.60 MK3 BAR / DOT OISPLAY

Displays an analogue voltage on a linear 10-element LED display as a bar or single dot I ideal for thermometers, level indicators, etc. May be stacked to obtain 20 to 100 element displays. Requires 5-20V supply.

stacked to obtain 20 to 100 element dispine). Requires 5-20V supply E4.75
MK4 PROPORTIONAL TEMPERATURE
CONTROLLER
Based on the TDA1024 Zero voltage switch, this list may be wired to form a burst fire power controller or a proportional temperature controller enabling the temperature of an enclosure to be maintained to within 0.5 C

3KW £5.55

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Based on the ZN 1034E Timer IC this kit will switch
a mains load on (or off) for a preset time from 20
minutes to 35 hours. Longer or shorter periods
may be realised by minor component changes
Maximum load 1KW

BOXES

Moulded in high impact ABS. Supplied with lids and screws. Black. B1 75×6×35mm B2 95×71×35mm B3 115×95×37mm



RESISTORS

ZENER DIODES

Pack of 10

80p

ARE BRAND NEW AND TO SPECIFICATION. ADD 30p PSP and 15% VAT TO TOTAL. OVERSEAS 50 (Europe), E4 (eistewhore) for PSP, Send sae for price list and with all enquiries, Callers welcome 0.00-4.00 (Stall).



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MAGNETIC FIELD
AMPLIFIERS

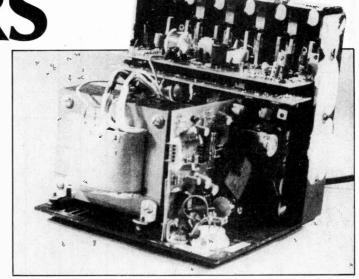
Once upon a time ETI ran an exclusive feature on magnetic field amplifiers. After nearly two years, this revolutionary design is finally in production, and it's time for another exclusive. Stan Curtis reports on the Carver Corporation's M-400.

or the past two years rumours have been rife about a radically new power amplifier conceived by Bob Carver, the designer of the original 'Phase-Linear' range of superamps. At last this amplifier, the M-400 'magnetic field amplifier' is in production and all has been revealed. Well, not quite all. In fact, to be honest, very little has been revealed at all. To keep the competition at bay a few red-herring explanations have appeared and even ETI's preview (July '79) showed a system of far greater complexity than has finally appeared in production.

So just what is it that is so unique about this new power amplifier? To answer that question properly, let's look at the Carver M-400 in detail. As the photographs show this amplifier is housed in a small cubical box (approximately 7 x 7 x 7 inches) which also serves as the heatsink. It is something of a lightweight weighing under 9 pounds and can be held comfortably in one hand. But then so can several of the Japanese 20 W (or so) 'micro-series' power amps. The difference is that the M-400 is rated at 200 W per channel and (into 8R loads) clips at nearly 300 W per channel! Yes, it does make you stop and think and if I hadn't had one to play with for a month or two, I might not have believed it either.

To make such a powerful amplifier so small, two major problems have to be solved. The first is minimising the amplifier's dissipation (ie heat), and the second is reducing the bulk of the power supply. Let us first look at the problem of amplifier dissipation.

If, as in the case of the M-400, a total supply voltage of 160 V is used and the output stage quiescent current is set at 100 mA, then the static dissipation will be equal to 160×0.1 , ie 16 W. For a stereo amplifier this would mean 32 W, making this particular magnetic cube far too hot to support in one hand! The solution is easy, as shown in Fig. 1, part of the circuit of one channel of the power amplifier. The output stage uses three pairs of complementary power transistors wired in series. Each pair is fed from a separate supply rail via blocking diodes. Low level signals are conducted to the load through transistors Q14 and Q15 which derive their power supply from the ± 25 V rails. Once the output signal increases beyond +23 V peak (driving an 8R load), transistor Q16 starts to conduct, drawing its current from the +48 V rail. The blocking diode prevents this higher voltage from sinking into the +25 V supply. Similarly for a -23 V signal transistor, Q17 will start to conduct. A quick



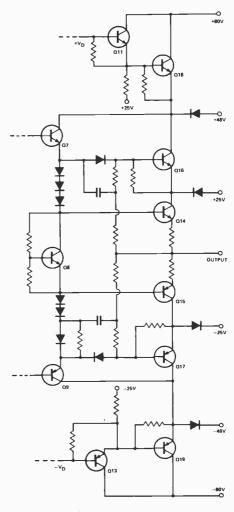


Fig. 1 Simplified circuit of the M-400 power amp output stage.

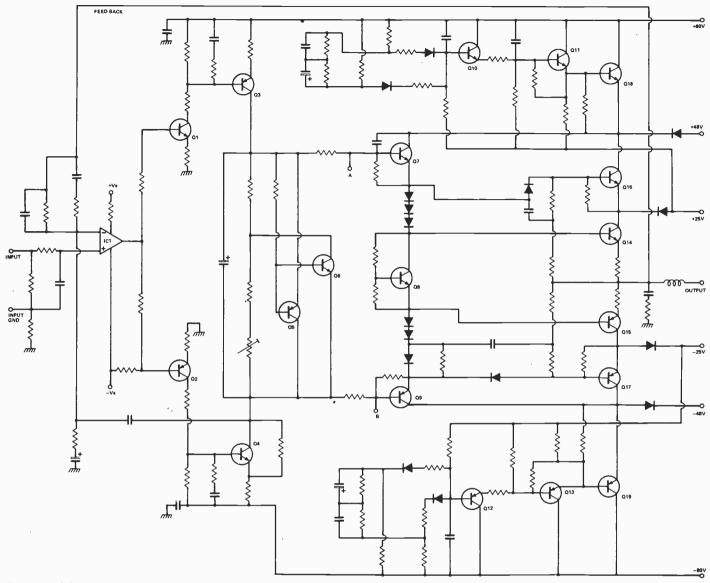


Fig. 2 Part of the circuit schematic of the Carver M-400 power amplifier.

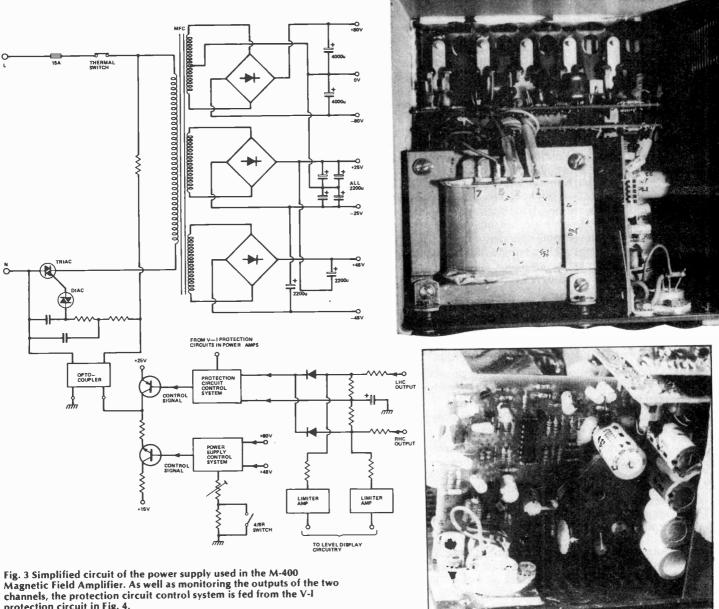
calculation will show that the level shifting diode chain creates an overlap situation, so that Q16 (for example) starts to conduct before Q14 runs out of voltage and causes clipping. This also avoids the occurrence of any 'crossover' type of discontinuity in the output signal. Although the first two pairs of output transistors (Q14 to Q17) are driven (and hence controlled by) the signal, the final pair (Q18, Q19) are just voltage sharing and are conducting at all times. Thus the current through these two transistors is controlled by the lower transistors, but only when the output signal exceeds about $\pm 47V4$ (peak). Obviously such an output stage permits the use of lower voltage (and less expensive) power transistors, but another major benefit is the much reduced quiescent dissipation of the output stage. Now remember that horrendous figure of 32 W? Repeating the same calculation with this type of output stage gives a dissipation equal to 48.8 x 0.1 or about 10 W total; in other words, reduced by a factor of three. That means a much cooler-running magnetic cube.

This output configuration is quite interesting but hardly radical, for similar arrangements have been used in other amplifier designs before the Carver M-400. Indeed, if the various supply voltages were obtained from a stable, high-quality, but

otherwise conventional power supply, the result would be very satisfactory but nothing out of the ordinary.

It is the power supply system of the M-400 that is radical, exciting and original. As everyone knows, the simplest conventional power supply consists of a power transformer, rectifiers and reservoir capacitors. These capacitors are used to store the energy in the power supply and thereby bridge the gap between the outflow of energy to the load (loudspeaker) and the inflow of rectified energy every 100 ms. Big amplifiers need to have very large reservoir capacitors with a ripple-rating that enables them to charge and discharge many Joules of energy. The laws of physics also dictate that with present day standards of metallurgical knowledge, small amplifiers have small power transformers; large amplifiers have large power transformers; and 400 W amplifiers have bloody massive transformers! Sony and JBL (amongst others) solved the problems of bulk by the use of switched-mode power supplies. The incoming AC mains supply is rectified into raw DC, which is then chopped on and off at a frequency of 50 kHz or more. This reconstituted AC voltage is stepped-down by a transformer and then rectified conventionally to generate the final DC supply. Now even a 400 W 50 kHz transformer can be wound on a comparatively small fer-

FEATURE: Magnetic Field Amps



protection circuit in Fig. 4.

rite core, so the problem of bulk is solved. But in its place come the problems of cost, circuit complexity, reliability and the generation of massive amounts of spurious signal radiation. Remember that the raw DC has a voltage of around 340 V, which spells expensive power transistors in any language! Suffice it to say that switched-mode power supplies for audio amplifiers have not caught on.

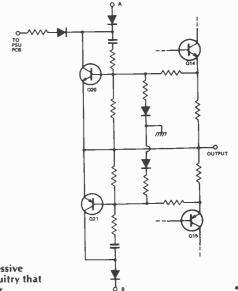
Carver has adopted a new approach. The first targets for elimination were the massive power transformer and the bulky (and expensive) reservoir capacitors. A glance at the circuit of the M-400 power supply (Fig. 3), and the photographs, will show that these components have not been completely eliminated but have been reduced in size and cost. The main reservoir capacitors have a value of 4000uF against the 15000uF which is typical of this class of amplifier. The transformer has been reduced to about one-fifth of its usual size. But then it isn't really a transformer because, despite appearances, the core and windings have been arranged to produce a saturable coil known as a 'magnetic field core' (MFC). This inductor is used to store energy for very short periods, thereby fulfilling one of the roles of the reservoir capacitors. Whenever the triac is turned on, a short pulse of current is fed into the MFC and builds up a

Top: Inside the cube. Looking like an ordinary transformer, the Magnetic Field Core can be seen at the bottom of the box. Above: Close-up of the power supply control board. The radical design means that many of the reservoir capacitors

can have values as

low as 2200uF.

Fig. 4 (left) Output stage V-I protection circuit, just part of the impressive range of sensing circuitry that protects the amplifier.



FEATURE : Magnetic Field Amps

magnetic field. At the end of the current pulse the magnetic field starts to collapse, generating a current in each winding. The 'primary' winding is effectively open-circuit because the triac is 'off' so the current is concentrated in the 'secondary' windings. This current will seek the lowest impedance path ie the discharged reservoir capacitors. So, of the three power-supply rails only those delivering current to the load will demand current from the MFC. Thus, in part, the power supply 'tracks' the demands of the audio signal.

The whole action is broadly similar in concept to the ignition-coil circuit in a car except for the voltages involved. If the triac is driven by a circuit which monitors the final DC voltages, the result would be a stable regulated supply. By using such a form of regulation in the M-400 the makers ensure that the power supply always responds to the current demand of the power amplifiers. It works well, too. My measurements showed that the supply rail voltages only varied by a few volts between no load and full loading (ie clipping into 8R loads). This novel power-supply arrangement has enabled Carver to include many extra protection circuits in the M-400. The circuit of the power amplifier shows that the normal V-I protection arrangement is used to clamp the signal level at the bases of the driver transistors. But, in addition, this protection circuit is coupled to the power supply control board. Also on this board are;

- i) an output DC-offset voltage sensing circuit
- ii) a clipping detector which senses the presence of high frequency components generated by sustained clipping of the output signal
- iii) a voice coil integrator which monitors the long-term average power fed to the speakers
- iv) a differential low-frequency circuit which monitors the two outputs. Normally the very low-frequency signals (of a music signal) are in mono, ie both channels are in phase. But suppose that you drop the pick-up onto the record; a large vertical signal will be produced. This will appear as a large out-of-phase component and trip the protection circuit.

The outputs of all these protection circuits are summed and their operation causes the triac to be turned off. The power supply has a relatively short time constant, so within that time the amplifier runs out of power and is rendered 'safe'.

As protection systems go it is pretty effective and quite fast-acting. However, I found it to be a little too sensitive and something of a nuisance in operation. I dislike protection circuits which make it necessary for you to wait a few seconds (for the capacitors to discharge) before switching it back on.

Another interesting aspect of the power supply is the way the two power amplifiers are wired out-of-phase. This arrangement permits instant conversion of the amplifier into a mono amp with a bridged output. Yet stereo operation is still quite straightforward, and is achieved by reversing the polarity of one pair of the speaker terminals. More importantly, this means that the current flow to the amplifiers is out-of-phase. For a reasonably symmetrical signal, one channel will be drawing current from the positive supply while the other draws current from the negative supply. Thus there is little likelihood of dynamic crosstalk through the supply lines, and a smoothing out of the current demands from the supply.

So how does this new super-amp perform? You'll have to wait for your resident audiophile, the editor, (all kneel) to give his opinion about its sound quality, but I can tell you how well it measures. See Table 1.

As can be seen the results are quite good except for the high-frequency distortion, which is higher than in many other large amplifiers.

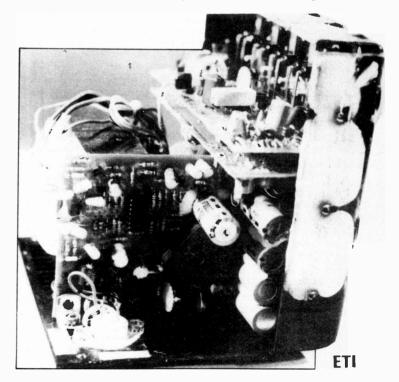
Power Output (20-20,000 Hz) 1% max. THD	300 W * 200 W (*protection lim	8R 4R nited)
Dynamic Power Output IHF toneburst	320 W 440 W 420 W	8R 4R 2R
Total Harmonic Distortion 200 W into 8R	0.01% 0.006% 0.07%	20 Hz 1 kHz 20 kHz
Sensitivity Input for 200 W into 8R	1.25 V	
Signal-to-Noise Ratio 'A-Weighted', 1 W into 8R	82 dB	
Frequency Response (small signal)	−1 dB −1 dB	3 Hz 53 kHz

Table 1. Performance of a typical Carver M-400 amplifier.

Of course the M-400 does have disadvantages. With all that current pushing and pulling in the coil there are some very strange noises emitted. The groans, squeaks, buzzes and the like can make you very nervous, but apparently it's quite normal and nothing to worry about. The M-400 can also develop quite an appetite for mains fuses. When it is seen that these have a value of 15 A it can make you wonder! The M-400 draws very high currents from the mains supply, but this current is out-of-phase with the voltage so at the end of the day the power (and hence the electricity bill) is comparatively low. Nevertheless, 15 A is 15 A and one can't help but wonder where it's all going.

These niggles apart, the M-400 is an interesting amplifier and all the signs are that other manufacturers will be beating a path to Carver's doorstep to discuss licence agreements.

Already I can see an immediate application in the design of high power in-car amplifiers. The words 'Magnetic Field' seem set to become an established part of audio terminology.



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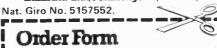
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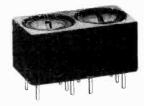
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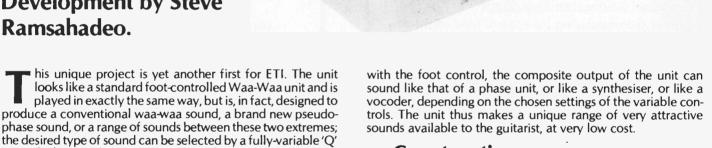


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WAA-PHASE UNIT

A superb project for the guitarist, this unit adds a brand new range of sounds to a player's repertoire. The design can produce standard waawaa or a unique pseudophase sound, has built-in mixer and balance controls, and costs a mere £15 or so to build. Design by Ray Marston. Development by Steve Ramsahadeo.



unit is switched in and out by a built-in bypass switch.

The completed project is housed inside a neat but robust foot-pedal unit that comes pre-drilled to accept all the control pots, switches and jack sockets that are used in the project, which is powered by a pair of PP3 batteries. We estimate the total building cost of the unit at about £15, including the price of

control. The unit also incorporates a built-in audio mixer, enabling the original guitar signal and the waa-phase signal to be

mixed in any desired ratio, and has a balance control so that no

apparent shift occurs in the guitar's mean sound level when the

the foot-pedal unit.

Basic Principles

A guitar produces output waveforms that are very rich in harmonics, which gives the instrument its characteristic sound. In a conventional waa-waa unit, the guitar output signal is simply passed through a foot-controlled band-pass filter before reaching the main amplifier. This filter is a low-Q type (typically with a Q of unity) and passes a broad spectrum of basic guitar sounds, but at the same time picks out and accentuates certain harmonics; when the operator sweeps the filter up and down manually using the foot pedal, the characteristic waa-waa sound is produced.

The unique feature of the ETI Waa-Phase unit is that its sweep filter has a Q that is fully variable from unity to eight. When the Q is set to unity, the circuit produces conventional waa-waa sounds. However, when the Q is set to maximum the filter picks out selected harmonics, amplifies them, and converts them to very pure tones that are quite unlike those of a normal guitar. These tones can be added to the original guitar signal via the built-in mixer. When the filter is swept manually

Construction

The foot-pedal unit used with this project (see Buylines) is supplied pre-drilled to accept all pots, switches and jack sockets that are used in the design, so construction should present very few problems.

Start by assembling the components on the PCB, as shown by the overlay, noting that Veropins are used to facilitate the interwiring to the rest of the circuit. When the board is complete, secure it in place inside the foot-pedal unit with a couple of sticky-pads, then proceed with the interwiring to the four pots, two switches, two jack sockets and the two batteries.

When interwiring, take extra care to conform to the circuit diagram. Note, for example, that the two halves of RV3 are contra-connected, so that the output of one half increases as the other decreases. Also note that the action of the foot-pedal unit is such that it sweeps only 200° or so of the available range of RV2, so position this pot carefully so that its value can be swept all the way from zero resistance to some high value.

When construction is complete, fix the two batteries in place (one in the built-in battery holder, the other secured with a sticky-pad) and give the unit a functional check. Simply connect the guitar output to the input of the unit, take the output of the unit to an amplifier, turn the unit on, switch SW1 to IN, and then

vary the controls while playing the guitar.

Start by setting the mix, balance and Q controls to mid value while you get the basic feel of the unit, then vary the Q and mix controls to explore the full sound range of the device. In final use, set the Q and mix controls to give the sound that you like best, then set the balance control so that negligible apparent change in sound levels occurs when the unit is switched in and out with SW1. The unit is then ready for stage use.

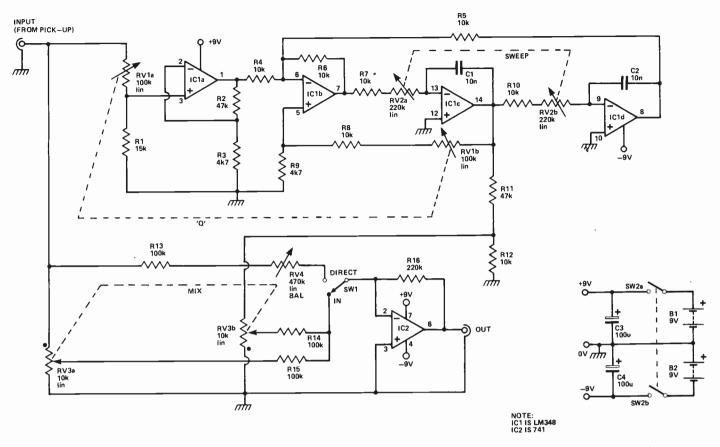
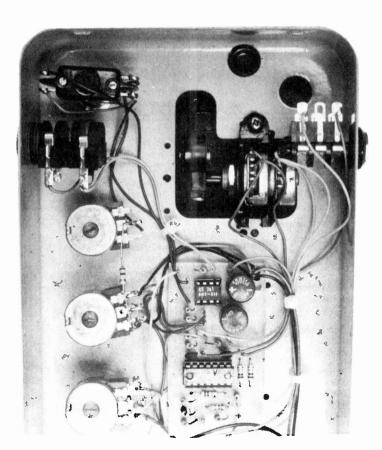


Fig. 1 Circuit diagram of the Waa-Phase unit. The dots at the ends of RV3a and RV3b indicate the left-hand terminal of the pot as seen from the rear. Make sure that this pot is wired correctly.



HOW IT WORKS.

The unit comprises a combined preamplifier and variable-Q, variable-frequency band-pass filter designed around IC1 (a quad

variable-frequency band-pass filter designed around IC1 (a quad op-amp), plus a simple two-input audio mixer designed around IC2. The guitar output signal is passed through the band-pass filter and the resulting signal is then mixed, in any desired ratio, with the original guitar signal, producing a composite output that can be fed to an external power amplifier.

IC1a acts as the preamplifier and drives the state-variable band-pass filter that is formed by IC1b-IC1c-IC1d and the associated components. The centre frequency of this filter is varied by foot-operated two-gang rheostat RV2. The Q of the filter can be varied from unity to eight by RV1b; to maintain an effectively fixed input-to-output gain at all Q settings, the gain of the preamp stage (IC1a) is varied in opposition to the Q by RV1a, the other half of the Q-control rheostat. other half of the Q-control rheostat.

The output of the filter (taken from the R11-R12 junction) is ine output of the filter (taken from the K11-K12 junction) is mixed (added) with the original guitar output signal by the network consisting of RV3-IC2 and associated components. Note here that the two halves of mix pot RV3 are contra-connected, enabling the final output signal to be varied from 'all guitar' to 'all waa-phase' (or any desired mixture) using the single control.

When SW1 is switched from the IN to the DIRECT mode, belong control RV4 can be directed to give no apparent change in

balance control RV4 can be adjusted to give no apparent change in the mean acoustic output levels between the two modes.

Close-up of the important bits. This photograph shows how R13 is mounted on the potentiometers. Note that the recommended case (see Buylines) comes with all the necessary holes ready-drilled. There is just sufficient room in the foot-pedal cut-out to accommodate RV2, a dualgang pot; connecting the latter to the pedal linkage is quite straightforward.

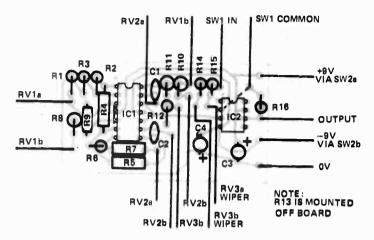


Fig. 2 Component overlay. Don't forget the insulated link between IC2 pin 2 and R16.

BUYLINES.

The neat pedal that houses the Waa-Phase circuit is available from Sola Sound Ltd. The pedal gives a professional finish at just half the price of most commercial units. It comes pre-drilled to accept all the panel hardware and is priced at £7.48 including VAT. Order as WFS1.

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R2,11 47k
R3,9 4k7
R4,5,6,7,8,10,12 10k
R13,14,15 100k
R16 220k

Potentiometers

RV1 100k linear, dual gang RV2 220k linear, dual gang RV3 10k linear, dual gang RV4 470k linear

Capacitors

C1,2 10n ceramic C3,4 100u 16 V el

100u 16 V electrolytic, PCB-mounting

Semiconductors

IC1 LM348 IC2 741

Miscellaneous

SW1 SPDT footswitch SW2 DPDT miniature toggle

PP3 batteries (2 off), jack sockets (2 off), waa-waa pedal (see

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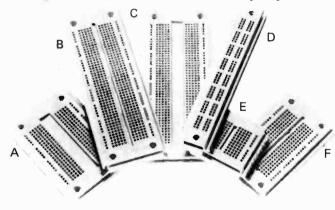
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AUDIOPHILE

This month Ron Harris looks at (and listens to) a new tuner from those awfully nice Sony people. For moving coil buffs we have the MC30 and a couple of step-up transformers, and there's a mouth-watering sneak preview.

B efore starting on what is, let us consider what shall be. Next month in ETI we are publishing an audio amplifier design for the home constructor. It is easily constructed and set up, has a preamp of comparable quality to ANY commercial unit and produces 150 W of good clean power into a real load.

Nothing too unusual there, you may say — typical ETI genius, you might add. However, there is one more little thing you ought to know before you turn the page, put the mag down, or pick up the knitting again. This amplifier is CLASS A. That's right — the one with all the heat!

Physically the System-A, as we're calling it, comes in four cases. Two mono power amps, a preamp with modular gain blocks and a PSU for the low-level modules. Future expansion plans include a two/four channel parametric equaliser for room correction, and a bridging module to allow two System-A power

amps to be used per channel, with massive power gain.

As the preamp output impedance is low, long runs of screened cable to the power amps present no problem — thus the large black boxes can be hidden from the view of wives, interior designers, budgies — or anyone else who might fall over them.

The factor that makes the System-A so different is the quality of its sound output — it is truly outstanding. It lies firmly in the highest possible class and will better most amplifiers on sale today, regardless of price. If anyone out there is thinking of buying, building or stealing an amplifier of between 60 W and 120 W and costing between £100 and ∞ — wait. For the sake of your ears — wait!

This is something really special and from now on will be the standard against which all else is judged. Don't miss out on this one — better to go deaf!

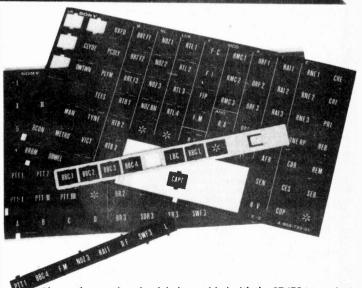


Sony ST-J75 Tuner

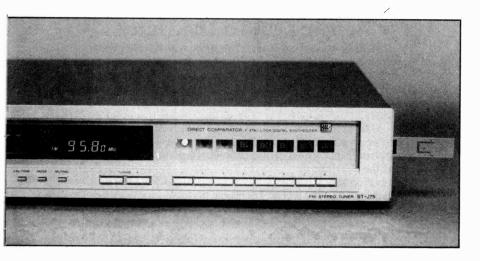
Alright, so I'm doing a tuner, now will you stop writing in asking what's happened to radio on these pages? The simple fact is that FM reception has changed but little in recent years. All the 'flagship' tuners comfortably exceed broadcast quality on most counts and differences between well-designed tuners are small compared to those encountered between other components in the hi-fi chain.

No doubt I shall now be castigated in some quarters (makes your eyes water that does, I can tell you) for that generalisation. You can annoy all of the people all of the time, but you can only please some of the people none of the time. Something like that anyway

This new Sony model, I feel, is a significant step forward and as such deserves recognition. Technically it is a MOSFET design, using a phase locked loop (PLL) decoder and copious filtering to ensure good linearity and outstanding spurious rejection. The PLL uses a very high comparison frequency (50 kHz), eliminating the need for a prescaler in the divider circuitry. The signal-to-noise ratio benefits greatly from this, achieving a figure of 80 dB (stereo!). Other specs. can be obtained from Table 1. Note that this is an FM-only unit.



Above: the set of station labels provided with the ST-J75 tuner. Just about every radio outfit in Europe gets a mention here somewhere. Neat idea this.



Left: how those same labels are operated. The plastic slide is removed from the side of the front panel, and LEDs behind them provide the illumination. The space around the station frequency on the main display is utilised by a "program" counter which relays the information as to which part of the pre-set list the tuner has reached.

Digits Definitely

The ST-J75 is a true digital design. There is no tuning control in the conventional sense at all. Instead there are eight preset stations and up/down counter control of tuned frequency, plus muting and mode commands, so that when retuned weak stations reappear in mono with low muting level.

In addition there is a program control, which can be used in conjunction with the preset channels and an external timer to record from each of the stations in any predetermined order! This operates using the tuner's non-volatile memory to store a sequence and will 'clock along' each time the unit is reenergised by the timer unit. Clever, that.

The 'manual' tuning buttons will step the frequency by 0.05 MHz on each depression, or rapidly run up or down if held on. In addition a 'scan' facility will check all the presets in order. There are indicators for all the facility 'states', the mode switch doubling for mono or stereo. This is useful for low-level stations which are strong enough to push the tuner into stereo, but too weak to give a good signal-to-noise ratio.

A 'calibration tone' generator is included which produces a 400 Hz tone at a level roughly equal to that produced by the tuner at 40 kHz deviation. This is used for setting up a tape deck to record a program at some later time, say with a timer. After a couple of experiments it would be possible to get very good recordings this way, once the deck is calibrated to the tone level.

I made several recordings from Radio 3 this way and obtained fair results first time and superb results third time! (We won't talk about the second time — I forgot to switch the timer on!).

Test Bench

I've been getting very fed up with some hi-fi recently. There is little point in spending my Sundays crouched in front of a bank of winking instruments, like some latter-day Quasimodo, if at the end of it all the blasted equipment exceeds spec. on all counts. Where are the drop-outs and deviations of old? Wither now the smoking ruins of output stages wrecked by 2uF at 10 kHz?

No wonder reviewers are spending a great deal of time and effort devising new tests — they're hoping something will FAIL! Note the capitals. Articles are more *fun* when things go wrong. Perfection can get a little boring, after all. Still, if I must...

The Sony ST-J75 acquitted itself impeccably under test and exceeded its specifications on all bar two counts. Differences in measurement techniques would account for those.

Damn it.

A Capital Sound?

On then to the living room, armed with yards of co-ax and phono leads. As I mentioned earlier I experimented with the program function and found it more useful than expected; it would be ideal for recording a series of concerts, for example.

The muting levels proved sensibly set, with the more vicious setting removing all but the strongest stereo signals. Tuning is precise and simple and no problems were encountered in use. After this, going back to a mere tuning knob would be like getting out of a Rolls and riding home on a bike!

The sound quality of the tuner is revealingly good. It carries an 'openness' to it that I have not heard before. Providing the programme is up to it, of course, the ST-J75 is capable of making sweet music indeed. When replaying live concerts the ambience of the hall was nicely apparent and the sound had none of the sense of restricted range which can beset lesser units at times. If only records sounded this good, there would be no need for digital recordings! Comparing it to my present reference — a Pioneer TX 9500 II — the Sony was shown to have a smoother response with greatly improved mid-range rendition.

Concluding Lines

At a typical retail price of under £200 (just) the ST-J75 is intended for the audiophile. However, judged against the facilities and sound quality, that price cannot be called less than good value. Anyone making off-air recordings would value the Sony highly.

Overall it produced better results from FM than any other tuner I have heard to date.

Table One — Test Results ST-J75

Tuning range: 87.5 MHz — 108 MHz

Sensitivity (26 dB quieting): 1.4 uV

Signal-to-noise ratio: 80 dB (stereo); 88 dB (mono) THD (1 kHz): 0.05% (stereo); 0.02% (mono) IM Distortion: 0.03% (stereo); 0.02% (mono)

Frequency response: 30 Hz-15 kHz ± 0.5 dB

Channel separation (1 kHz): 57 dB

Output level: 763 mV into 600R

Capture ratio: 1.2 dB Image rejection: 100 dB AM rejection: 68 dB

Price: £199 (typical)



MC30/T-30 — At Last!

After having been postponed for a month, the Ortofon MC30 makes its cantilevered way into ETIprint. The recommended step-up device for this low-output moving coil is the T-30 and I'll be taking a look at that, too.

Both these items are priced at over £200 and are strictly for the high of fi and low on money worries. However, they incorporate sufficient ingenuity as to render them highly interesting to read about (I hope).

As an interesting comparison to the T-30 I managed - at the last moment — to include a comparison with a lower priced transformer, Mayware's new T-24 II device. Unfortunately this was too last moment to get a full lab test, but I did manage to verify frequency response and gain some indication of phase shift. At £69 it makes an interesting alternative.

Ortofon Principles

Of the many refinements contained in the MC30 design. perhaps the most interesting is the damping system. The cutaway (below) gives an idea of the complexity involved.

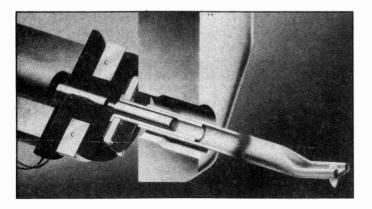
The system purports to optimise, across the entire audio band, the damping applied to the cantilever. At low frequencies the platinum disc will 'decouple' the two sections allowing for effectively reduced mass and improved tracking. Neat and effective — the MC30 is the best tracking moving coil cartridge I've encountered.

The cantilever is composed of an aluminium alloy which has low mass and great rigidity. The stylus is a 'fine line' design of low tip mass (0.4 mg) which helps extend the high frequency

tracking capabilities.

In addition the coils themselves are wound with very small cross-section wire. Each 'layer' contains only 20 turns of wire. Each MC30 is individually tested and calibrated and the results are provided to the purchaser — as are a test record and headshell

One drawback of the 'small coils' approach is that the MC30 has a lower output than is usual and thus requires greater gain in the step-up device used with it. This brings me nicely to the T-30. Ortofon's own required transformer.



Stepping Up To The Thirty

Transformers have fallen from grace as that with which to increase moving-coil signals — mainly due to the phase linearity problems inherent in poorly designed units. Solid state amps, such as Sony's excellent HA-55, offer a clean and extended response, albeit at a cost of some noise addition and PSU prob-

Transformers require no power supply and if the linearity problems can be overcome they offer a more elegant solution to the problem. Overcoming problems is generally expensive and the T-30 is no exception to this! Let me say now, though, that its performance is practically beyond belief for a transformer, regardless of what it cost. Technical performance would do any amplifier proud.

Toroidal coils are employed and wound so as to reduce phase-shift overall. The gain/input impedance is switchable from 3R(32 dB gain) to 48R(20 dB gain), and a bypass position is provided for ease of comparison with moving-magnet designs. Gold-plated connectors are used throughout.



The two transformers under consideration in this months article. The Ortofon boasts that input switching and a bypass facility. The Mayware boasts a much lower price!

Technical Tests

Not twice in a month I don't. Full test results are given in the tables, and comparisons with the manufacturer's specs will 'reveal absolutely nothing! The T-30 has a better square wave performance than many amps I've tested and phase deviation was always within 12° (5 Hz-50 kHz) ie not worth worrying about.

The Mayware T-24 II was not given the full treatment there wasn't time — but the tests I did came out very well indeed and boded well for the auditioning to come later.

Down To The Sound

Using a cartridge like the MC30 can be most disconcerting at first. Like all true hi-fi it adds little of itself to any material passed through it. This means that good records sound superb. It also means poor records sound rotten! Honesty is always preferable in the long run, and the MC30 scores highly on that count.

To avoid repeating superlatives I shall simply say that there is little, if anything, to criticise in the MC30 sound. It is neutral, clear and detailed. Bass response is firm and extended with no hint of boom. Treble is sharp without being hard and the midrange beautifully defined. It is without doubt the best pick-up cartridge around, in my opinion, and will remain my point of reference for some time to come, I think.

Left: the inside story on the Ortofon wide range damping system. Note the size of the coils and the position of the platinum disc (marked "c") referred to in text. This is to decouple the system at low frequencies, being the cantilever mass.

If you are in the market for an ultra-fi cartridge don't miss the MC30 off your list. It may be fashionable to pay £500 for Oriental designs, but I fancy that £300 of that could be better employed. If the rest of your system is up to it, then this Ortofon will be money well spent.

Transformer Changes

Experimenting with the two step-up devices proved interesting. Both had an improved clarity over a couple of (under £100) transistor head-amps | compared them to, with far fewer worries as to hum loops and signal-to-noise.

The T-30 has as brilliant an audible performance as it is technically perfect. It has the edge over most head-amps in that it demonstrates an improved attack and handles the leading edges of sounds much better.

The Mayware T-24 II proved to be excellent value for its £69 price tag. It clearly out-performed head-amps costing much more, especially in the mid-range. The bass is well controlled, although maybe not quite as extended as the T-30. The two were distinguishable on audition, mainly due to that difference in bass response, but not as much as the price tags would suggest. The T-24 has no switched inputs, but will match practically any cartridge. I tried with the MC30, a Coral MC81 and an Entre 1, good results being returned with all.

The two units cannot really be considered competitors there is around £150 difference in their price! Within their own. realms both are outstanding and the T-24 can be confidently recommended for those beginning to explore moving coils. The T-30 is for the more experienced — and the more pecunious and will guarantee excellent results, albeit at a high price.

Table Two — Test Results MC30

Compliance (vertical and horizontal): 13 x 10⁻⁶ cm/dyne (cu) Output level (1 kHz, 5 cm): 0.08 mV

'fine-line' Stylus type: Separation (15 kHz): 22 dB Frequency response (20 Hz-20 kHz): $\pm 0.7 dB$

Optimum tracking weight: 1.7 g Weight: Typical price:

Left: the MC30 and the SME snaking their way into text. That cartridge body protrudes considerably does it not? Still looks rather stylish though. Right: Internal details of the T-30 step-up. Not a lot to see in here is there? All the connectors are gold plated (naturally) and the screening is superb!

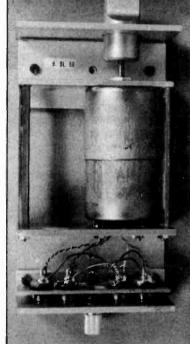


Table Three — Test Result T-30 and T-24 II

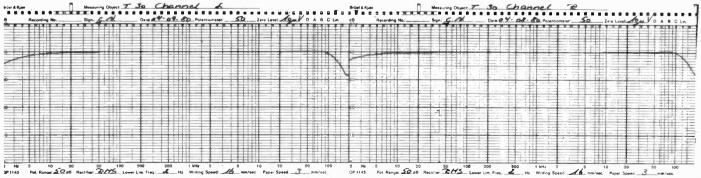
T-24 Input impedance: Output impedance: 47k; 150pF

Frequency response: $5 \text{ Hz-}100 \text{ kHz} \pm 1 \text{ dB}$ $20 \text{ Hz-}20 \text{ kHz} \pm 1.5 \text{ dB}$ Gain: 22 dB-33 dB dependent upon Z

Separation (1 kHz): 55 dB \pm 12° (5 Hz-50 kHz) Phase linearity: 17º max Balance: within 1 dB within 1 dB

Note: T-24 results are not as comprehensive due to time pressures. Within the audio band its performance was comparable to the T-30, if not quite equal in all respects.

Last months review of the Mayware MC-3L Cartridge gave the price as being around £53. It was wrong. Actual price will be £49-ish and I must confess that I have no idea where I got £53 from, now you mention it. Put it down to too much claret and I shall accept the slapped wrist as gracefully as I am able.



E INSERE

Above: frequency response of the T-30 transformer. It has absolutely no right to be able to perform like this — many amplifiers cannot do as well!

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MICROCOMPUTER JOYSTICK CONTROLS

Video games have never been more popular, but many require the use of joystick controls. This A-to-D converter, submitted by I. Forster of Chelmsford, will provide such controls for any microcomputer using a 6502, and has many other applications.

his article describes a system for analogue-to-digital (A-to-D) conversion, originally intended to provide cheap and simple joystick controls for a Commodore PET computer. The system should work on any computer having a USER port, although the software necessary will probably be different to that used by the PET. The software is in 6502 machine code and can be merged with a BASIC program or used as a machine code subroutine.

Apart from joystick controls there are a number of other possibilities inherent in the principle. A few of these are described, although no practical results have been obtained with these circuits yet. The field is open for the experimenter!

There are a number of improvements possible, such as using both the X and Y registers in the 6502 as counters, giving 64K resolution on an input. Hardware improvements are also very definitely possible, although beyond a certain point it would probably be much cheaper to build a dedicated device.

Joystick Controls

The circuit for use with a joystick unit is shown in Fig. 1. Two of these are necessary, one for the X axis, one for the Y. The flow diagram for operation is shown in Fig. 2. X volts is the transfer voltage of one gate in the 4081 BE. This is not very predictable and better results could be obtained using a Schmitt trigger of some kind, such as the op-amp circuit shown in Fig. 3 or a CMOS Schmitt gate. If C1 is increased, keeping the charging resistor constant, the time taken to reach X volts becomes larger. If C1 is made too large the counter (in this case the X register of the 6502) will overflow and count round.

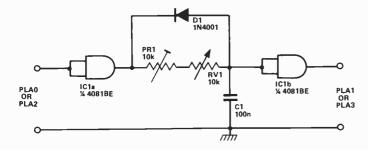


Fig. 1 Simple joystick version of the analogue-to-digital converter.

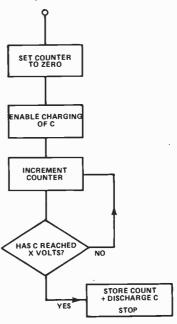


Fig. 2 Flow diagram for the ADC.

The software is disassembled Hex machine code, as shown below. It could be entered via the PET TIM monitor or as a data statement in a BASIC program. From BASIC it is called by SYS 826 and the values of the two conversions are stored in 1022(X) and 1023(Y); they can be retrieved by a PEEK instruction

LOC'N	CODE	INSTRUCTION
033A	78	SEI
033B	A9 55	LDA *55
033D	8D 43 E8	STA E843
0340	A9 00	LDA *00
0342	8D 4F E8	STA E84F
0345	A9 01	LDA *01
0347	8D 4F E8	STA E84F
034A	A2 00	LDX .00
Ó34C	E8	INX
034D	AD4F E8	LDA E84F
0350	C9 A3	CMP *A3
0352	D0 F8	BNE 034C
0354	A9 00	LDA *00
0356	8D 4F E8	STA E84F

0359	8E FE 03	STX 03FE
035C	E8	INX
035D	EO FF	CPX *FF
035F	D0 FB	BNE 035C
0361	A9 55	LDA *55
0363	8D 43 E8	STA E843
0366	A9 00 .	LDA *00
0368	8D 4F E8	STA E84F
036B	A9 04	LDA *04
036D	8D 4F E8	STA E84F
0370	A2 00	LDX .00
0372	E8	INX
0373	AD4F E8	LDA E84F
0376	C9 AC	CMP *AC
0378	D0 F8	BNE 0372
037A	A9 00	LDA *00
037C	8D 4F E8	STA E84F
037F	8E FF 03	STX 03FF
0382	E8	INX
0383	EO FF	CPX *FF
0385	D0 FB	BNE 0382
0387	58	CLI
0388	60	RTS

Simple Sums

The calculations involved are simple and are shown below: Fig. 1 refers.

Let X be the transfer voltage of IC1b — ie the voltage at which IC1b's output switches high. Then the time taken to get to X volts can be derived thus:-

so
$$t = \frac{C1.V}{I}$$

Since V = X(transfer voltage)

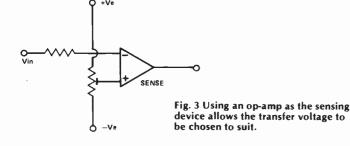
$$t = \frac{C1.X}{I}$$

But I =
$$\frac{V_{DD}}{R}$$

where R depends on the setting of PR1 and RV1

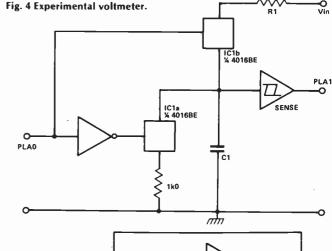
So t =
$$\frac{\text{C1.XR}}{\text{V}_{\text{DD}}}$$

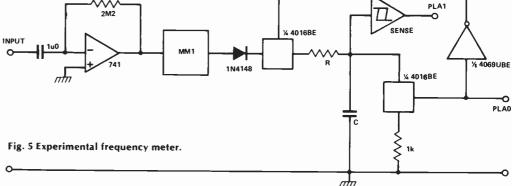
C1, X, V_{DD} are constant and so t is proportional to R.



Time And Again

Two other circuits are shown in Fig. 4 and Fig. 5 which use the principle of measuring the time taken to charge a capacitor. In both cases the symbol labelled 'sense' is a Schmitt trigger of some kind. The voltage across the analogue gates of the 4016BE should be less than the power supply to the chip. However, a simple op amp prescaler theoretically allows voltages up to the breakdown voltage of the resistors used to be measured. All the levels returned to the PET must be 5 V logic, so the 4016, inverter and sense output should all be 5 V logic. IC1b in Fig. 4 could be removed, although this complicates calculations because the capacitor will not start charging from 0 V.





The circuit of Fig. 5 is designed as a frequency meter with a range of 0-100 kHz. The 741 produces a square wave of the same frequency as the input. MM1 is a positive-edge-triggered monostable with a period of 10 uS; its output is fed via a diode, analogue gate and resistor to charge the capacitor. The leakage of the capacitor could be very significant so care must be used in selecting the values of R and C as well as component types.

Current can be measured by using a virtual earth type circuit for an op-amp. With a 741 only about 10 mA can be measured since it can only sink a maximum of 20 mA. A resistive divider network or a power op amp could improve this.

Conclusion

The circuits described are fairly crude but could offer usable results to the amateur electronics experimenter. Expense is a major factor in most circuits, and most of these could be built for a few pounds. Of course, you need a computer! Beware of locking your machine into endless loops — this is harmless to the computer but very wearing on your nerves. If not already provided, setting up a way of using the NMI (non-maskable interrupt) on your processor might be advisable. Good luck!

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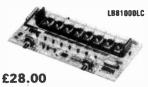


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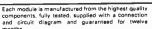
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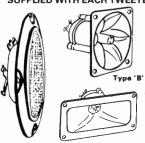
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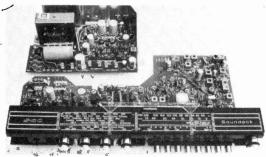
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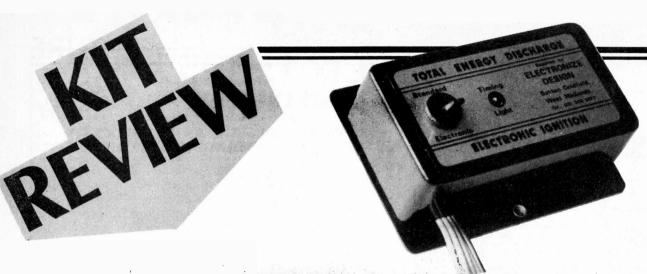
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lectronic ignition systems are quite popular these days and there are a number of models to choose from, both ready-built and in kit form. This month Kit Review examines one of the latest systems, the Total Energy Discharge ignition kit from Electronize Design. This patented design has an impressive pedigree, being the result of over 10 years' experience with ignition systems, and is currently being sold under licence by a major accessory manufacturer as a ready-built unit.

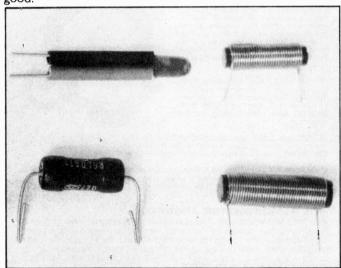
Ordinary discharge systems produce a high power spark but the very short spark duration makes them incompatible with the weak air/fuel mixtures used for economy in modern cars. Inductive systems are cheaper and provide a low power, long duration spark to guarantee ignition of the fuel, but the actual firing point varies a great deal from cycle to cycle and reduces engine efficiency. Electronize claim that their Total Energy Discharge system with its very high power, medium duration spark, will ignite the weakest of mixtures with the minimum of timing delay. The unit will also cope with low battery voltage and fouled plugs, has an LED for static timing, and a built-in switch allows you to change back to the standard system in an emergency. It is only suitable for negative earth vehicles.

What You Get...

Well, it all sounds good but any kit stands or falls by what you actually get for your money. How does this one measure up? The kit arrives in a large plastic bag; the bigger items are loose and the smaller components are supplied in several smaller bags. Checking the components off against the parts list showed that everything was present and correct, and in this case 'everything' means just that, right down to the last nut, bolt and connector. There's even a length of solder and a tiny tube of heatsink compound for the power transistor. All you have to provide are the tools and the car.

The PCB is good quality fibreglass with tinned tracks, and the holes are all pre-drilled to the correct sizes. Everything mounts on the PCB, including the transformer, switch and power transformer, and there are lots of little touches showing that someone has actually been thinking. For example, there is a wirewound resistor that dissipates considerable heat and needs to be mounted clear of the PCB to allow air flow. Electronize don't just tell you that this is necessary — they've bent the resistor leads double and crimped them so that the component is automatically positioned at the right height above the board. The LED has to be 23 mm above the board and the leads aren't long enough so it's supplied with extension wires already soldered on; furthermore, you don't have to worry which lead is

the cathode because colour-coded plastic sleeving has been fitted. The inductors are pre-wound with the enamelled copper wire glued to the formers, and the ends of the wire have had the enamel scraped off ready for soldering. This is the sort of attention to detail which separates the excellent kits from the merely good.

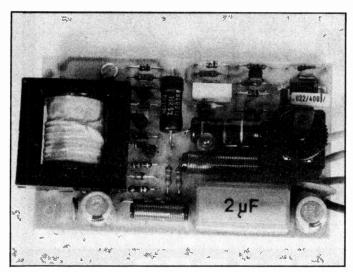


This is the way to supply kit components. The LED has colour-coded sleeving and the wirewound resistor is supplied with preformed leads so that it will mount at the correct height. The inductors are prewound, too.

The instructions are of a similar high standard. First there is a short explanation of how to solder, which should prove adequate for most beginners, followed by the actual assembly instructions. One of my criteria for the 'perfect kit' is that the instructions should answer any question that might occur to a constructor; except for one instance mentioned later, Electronize appear to have managed this. The assembly sequence is explained almost component-by-component. Where orientation is important the reference markings are fully described; in the case of the transformer orientation is not important, and you are told so. Anything that a beginner might do wrong seems to have been anticipated and warned against.

. . . And What You Do With It

Having been suitably impressed, it was time to get out the soldering iron and get on with the construction. The markings or colour-code of every single component is given in the parts list,



The completed board, ready to be fitted into the case.

Transformer

Trans

This is a copy of the component overlay from the instruction sheet. As you can see, it's very easy to follow and shouldn't cause problems even if you're a beginner.

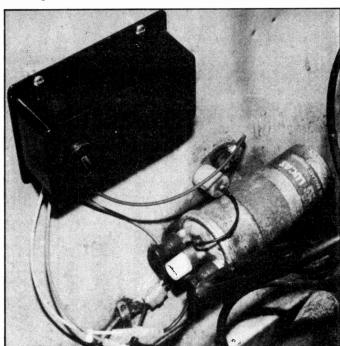
and for those items where, presumably, Electronize buy from several distributors, all the possible variations are listed. The overlay is very clear and easy to follow and assembly of the PCB was quite straightforward, except for a couple of points. The transformer and switch supplied in my kit were a very tight fit and I'd recommend that you test-fit these components prior to assembly in order to loosen up the holes. Second, when everything had been soldered in place there were several spare holes left on the PCB. Presumably the fact that some components may be supplied from a variety of sources means that the extra holes are to accommodate different case sizes nevertheless, this could cause a few nagging doubts for some people and it ought to have been mentioned in the instructions. The sheer pettiness of these criticisms can probably be taken as a compliment to Electronize — but I have to find something to complain about or the more cynical amongst you will think the kit came wrapped in used fivers.

The completed PCB was lacquered on both sides as recommended in the instructions (engine compartments not being the kindest of environments), and fitted into the case when dry. One end of the board is secured by two bolts (one of these also holds the power transistor against the mounting plate, which doubles as a heatsink); the other end is supported once the switch is attached to the plastic case. Make sure the LED and switch are passing through their respective holes in the case as you tighten up the screws — a moment's carelessness on my part resulted in a cracking sound and great panic! No harm done, fortunately.

Kit Fitting

Not having a car of my own, I persuaded my father, trusting soul that he is, to let me fit the unit to his Ford Escort. This, too, was a simple procedure and showed once again that the designer of the kit is on the ball. The ¼" bosses around the fixing holes allow clearance for the PCB mounting hardware — but they also make it easier to fasten the flat mounting plate to curving bodywork. This is not an important point on the Escort, which has a flat area conveniently close to the ignition coil, but it could make life easier on other makes of car.

Once the case was fastened in place the electrical connections were completed — all the necessary connectors are supplied. The power supply connections are made using a 'tap-in' connector to the ignition coil supply and an earth tag to a convenient point on the chassis; the two remaining leads allow the contact-breaker side of the coil to be connected via the electronic ignition unit. That's all there is to it!



The finished unit fastened inside the engine compartment of a Ford Escort. Only four electrical connections are necessary to make the unit functional.

Judgement Day

The unit worked first time, and works well — since it was fitted the engine has started instantly on every occasion, so I'm still welcome at home. The static timing light is a useful bonus and made it easy to check that the engine is correctly adjusted.

To sum up, the kit is very impressive, with a smart finish, well-written instructions and a good performance. It is obvious that a lot of care and thought has gone into it and at £14.85 it represents excellent value for money. Highly recommended.

PETER GREEN

BUYLINES

Total Energy Discharge Electronic Ignition system £14.85 including VAT and postage and packing. Electronize Design, 2 Hillside Road, Four Oaks, Sutton Coldfield, West Midlands B74 4DQ.

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57161	9.96	8224	2.50
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6502	7.96	8251 8253 8255	10.96
6520	4.50	8255	4.95
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LED JEWELLERY

Impress your girlfriends and test your building skills with these snazzy little flashing lights' jewellery projects. Design and development by Ray Marston.

he three jewellery projects described in the following pages can be worn as brooches by women, as badges or belt buckles by men, or as either by the in-betweens. Ideal for use at parties and discos, all three projects produce highly attractive visual displays and make delightful presents

The simplest of the projects has a four-LED display, is built on a single PCB and can be constructed in a hour or so. The second project has a 20-LED display, is built on a stacked pair of PCBs and should take several hours to build. The final project has a 25-LED two-colour display, is built on a stacked pair of PCBs and presents a real test of your constructional skills.

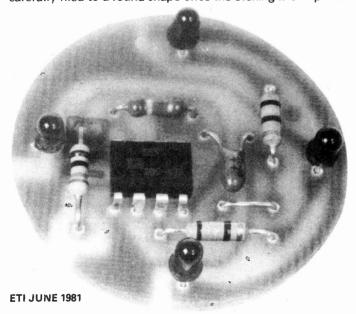
All three projects are designed to be powered by a PP3-type 9 V battery, hidden in the clothing of the user and connected to the jewellery by a pair of fine leads. The completed projects can be encapsulated in clear or tinted plastic resin and fixed in place with a clip bonded to the rear of the unit.

Cross Brooch

This simple little unit is built on a single circular PCB and has an overall diameter of 39 mm. The four-LED display is effectively arranged in the form of a crude cross and the circuit action is such that the vertical and horizontal bars turn on and off alternately, to produce an apparently moving display. The project requires a minimum of constructional skills and can be built in an hour or so.

Construction

The most difficult part of the construction of this unit relates to the manufacture of the actual PCB, which needs to be carefully filed to a round shape once the etching is complete.





The Star brooch, where Nature intended.

When the PCB is ready, assemble the components as shown on the overlay, taking care to test the four LEDs before fitting them into place. Note that IC1 is soldered directly to the PCB, without the use of a socket. When assembly is complete, give the unit a functional check by connecting a 9 V battery (PP3 type).

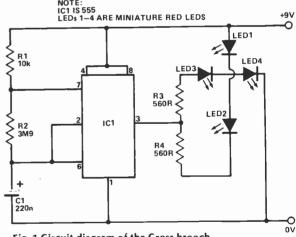


Fig. 1 Circuit diagram of the Cross brooch.

HOW IT WORKS

CROSS BROOCH

IC1 is a 555 chip, connected in the astable mode and oscillating at a rate of about 1 Hz. The output of IC1 (pin 3) is used to drive the four-LED display, which is arranged in the form of a cross. The four LEDs are effectively wired in series, with LEDs 1 and 2 and R4 in the 'upper' (vertical) half of the chain and LEDs 3 and 4 and R3 in the lower' (horizontal) half.

In each 555 operating cycle, when pin 3 of IC1 is low the lower half of the LED chain is shorted out and LEDs 1 and 2 illuminate via R4. When pin 3 of IC1 is high the upper half of the LED chain is shorted out and LEDs 3 and 4 illuminate via R3. The vertical and horizontal bars of the display are thus illuminated alternately.

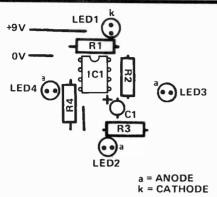


Fig. 2 Component overlay for the Cross brooch.

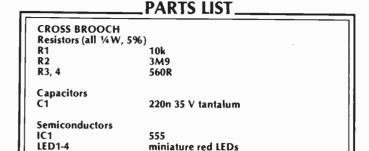
Spiral Brooch

The display of this 20-LED piece of electronic jewellery takes the form of five concentric circles of LEDs, with four LEDs in each circle: the LEDs in adjacent circles are offset from one another to form four distinct spirals. The action of the display is such that the circles illuminate sequentially, from the centre outwards, until the outer circle illuminates. At this point the display blanks for one-sixth of a second and then the entire cycle repeats. The effect of all this is that the display seems to repeatedly expand and twist, with a cyclic rate of about 1 Hz.

Construction

This unit is built up of two PCBs, one being used for the display and the other for the main circuitry. When construction is complete the two boards are fixed together using four wire struts. A good deal of care is needed in the construction of this unit; a fair degree of constructional skill is required and the total building time may be several hours.

Start the construction by etching the two PCBs and then very carefully cut the two boards to size. Now drill all the necessary holes in the PCBs, using a 1 mm drill and noting that the four strut holes are located very close to the edges of each



miniature red LEDs

Now proceed to solder the 20 LEDs on the display board. keeping the LEDs as close to the board as possible. Test each LED (using a 9 V battery and a 1k0 resistor) as soon as it is soldered into place.

Next, assemble the components on the main circuit board. Fit the wire link first, then the two resistors and two capacitors. then IC1, and finally IC2. Note that the two ICs are soldered directly to the PCB, without the use of sockets.

When construction is complete, interconnect the two boards noting that the display board connections are made to solder pads on the undersides of the PCB, and give the unit a functional check. If all is well, fix the two boards together (as close as possible) using tinned copper wire pushed through the strut holes and soldered into place. Note that the two PCBs are 'polarised' (a) by making the boards slightly non-symmetrical and (b) by the strut hole locations, to ensure correct alignment of the completed boards.

Fig. 4 Component overlay for

the Spiral brooch.

DCBA X

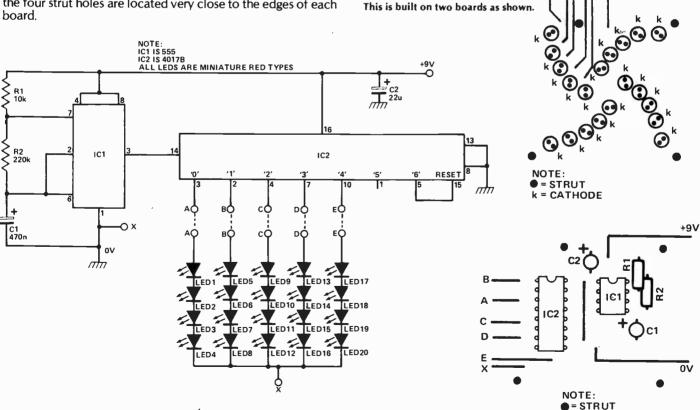


Fig. 3 Circuit diagram of the Spiral brooch.

PARTS LIST

SPIRAL BROOCH Resistors (all ¼ W, 5%)

R1 10k R2 220k

Capacitors

C1 470n 35 V tantalum C2 22u 35 V tantalum

Semiconductors

IC1 555 IC2 4017B

LED1-20 miniature red LEDs

HOW IT WORKS

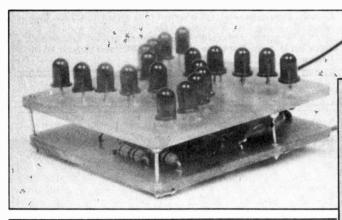
SPIRAL BROOCH

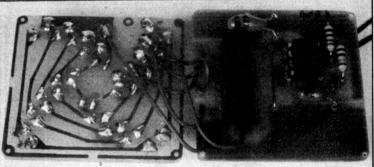
The electronic circuitry of this project is fairly simple. IC1 is a 555 chip, wired in the astable mode. This chip operates at a rate of about 6 Hz and feeds clock pulses to the input of IC2, a 40178 counter/decoder. The outputs of IC2 go high sequentially on the arrival of each pulse.

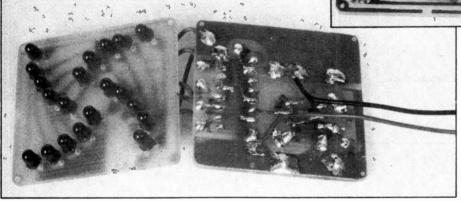
arrival of each new clock pulse.

The first five outputs of the 4017 are each used to drive a chain of four series-connected LEDs. Thus, on the arrival of the first five clock pulses in each sequence the chains of LEDs illuminate sequentially. On the arrival of the sixth pulse, no LEDs are illuminated. On the arrival of the next pulse pin 5 of IC2 goes momentarily high and resets the counter. This causes the '0' decoded output to go high, thereby initiating a new sequence of operations.

In practice, the LEDs of the display unit are arranged in the form of five concentric circles of LEDs, with four LEDs in each circle and with the LEDs in adjacent circles slightly offset from one another, to form four distinct spirals. Consequently, the display produces the expanding, twisting 'spiral' motion already described in the text.







These three photographs show the constructional details of the Spiral brooch. Two PCBs are used, with the LEDs soldered to one board and the remaining components on the other. The six interconnections are made by soldering wires directly to the copper tracks on the display board, then the boards are fastened together with short lengths of tinned copper wire. For compactness the capacitors are bent over sideways.

Star Brooch

The display in this 25-LED two-colour project is arranged in the form of a five-armed star (see circuit diagram), with four red LEDs in each arm, and with a single green LED placed between each pair of arms. The display is activated in a 10-step two-speed sequence. In the first five steps of the sequence, the five red arms of the star are sequentially switched on and off at a fairly slow rate, with the lower left arm turning on first and the lower right arm last. This action is followed by a rapid five-step sequence in which the three green 'G' LEDs turn on first, followed alternatively by red arms 'D', 'C' and 'B', with the three green 'F' LEDs turning on last. The sequence is then complete and immediately repeats, taking roughly 5 s to complete each 10-step sequence.

Construction

This unit is built up on two PCBs, one being used for the display and the other for the main circuitry. The construction of the unit needs much care and a fairly high degree of constructional skill.

Start the construction by etching the two PCBs and then carefully cut and file the boards to size. Now drill all necessary holes in the PCBs, using a 1 mm drill, noting that the unused holes shown in the photographs of our prototype display board are, in fact, erroneous.

Now proceed with the assembly of the 25 LEDs on the display board, keeping the LEDs as close to the board as possible. Once again test each LED (using a 9 V battery and 1k0 series resistor) as soon as it is soldered into place.

Next, assemble the components on the main circuit board. Fit the wire link first, then the resistors, capacitors, diodes and the three small transistors. Finally, solder IC1 and IC2 into place directly on the PCB.

When construction is complete, interconnect the two boards noting that the display board connections are made to solder pads on the underside of the PCB, and paying particular attention to the final orientation (top-to-top) of the two boards. Then give the unit a functional check. If all is well, fix the two boards together (as close as possible) using an epoxy resin adhesive, taking care to ensure that the boards are correctly aligned.

Finishing Off

Each of the completed brooches can be encapsulated in plastic resin (see Buylines), and fixed in place with a brooch clip bonded to the rear of the unit. Connections to the concealed battery can be made unobtrusively via a sub-miniature plug and socket of the type used for radio control servos.

BUYLINES.

Absolutely no problems here, as all components are readily available types. It is important to note, however, that all LEDs used in these projects must be good quality (low ON voltage) types: so-called 'bargain pack' LEDs must not be used.

The plastic resin for encapsulating the finished PCBs is sold as Plasticraft kits, which should be available from your local arts and

crafts centre.

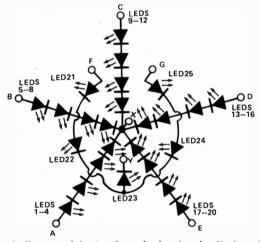


Fig. 5 Circuit diagram of the Star brooch, showing the display (above) and the drive circuitry (below).

HOW IT WORKS

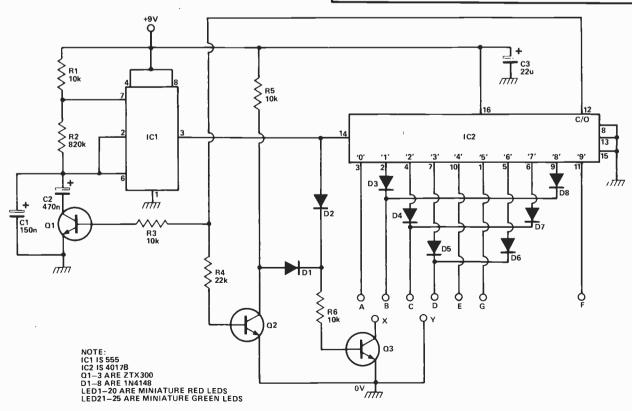
STAR BROOCH

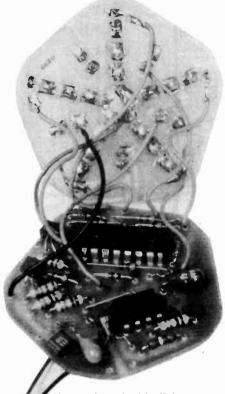
The circuitry of this piece of electronic jewellery is moderately complex. Here, IC1 is a 555 astable and feeds clock pulses to the input of IC2, a 4017B counter decoder which is wired in the divide-by-ten mode. The carry out terminal of this chip (pin 12) is high for the first five clock pulses of each 10-pulse cycle and is used, in our application, to control the speed of IC1 and the logic drive to the 25-LED display.

Thus, at the start of each operating sequence the high CO output of IC2 drives Q1 on and makes the IC1 astable operate at a slow rate, with a period determined by the combined values of C1 and C2. In the first five clock cycles outputs 'A' to 'E' of IC2 are driven high one after the other and the respective LED display arms are activated for half a clock cycle each by D2 and Q3. On the arrival of the next clock pulse the CO output of IC2 goes low, so Q1 turns off and IC1 operates at a rapid rate, with a period determined by C1 only. Simultaneously, Q3 is turned fully on via R5-D1-R6. Consequently, in the final five steps of the 10-step cycle, outputs 'G', 'D', 'C', 'B' and 'F' are rapidly sequenced on and off for one full clock cycle each. The cycle is then complete and repeats ad infinitum.

PARTS LIST

STAR BROOCH		
Resistors (all 1/41		
R1,3,5,6	10k	
R2	820k	
R4	22k	
Capacitors		
C1	150n 35 V tantalum	
C2	470n 35 V tantalum	
C3	22u 35V tantalum	
Semiconductors		
IC1	555	
IC2	4017B	
Q1,2,3	ZTX300	
D1-8	1N4148	
LED1-20	miniature red LEDs	
LED21-25	miniature green LEDs	





The Star brooch with all the interwiring completed and ready for encapsulation.

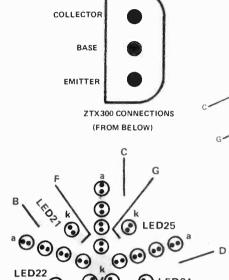
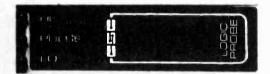


Fig. 6 Component overlay for both the boards of the Star brooch. The PCBs have an unusual shape so take extra care to orientate the polarised components correctly.

NOTE: LEDs 21 TO 25 ARE GREEN a = ANODE k = CATHODE

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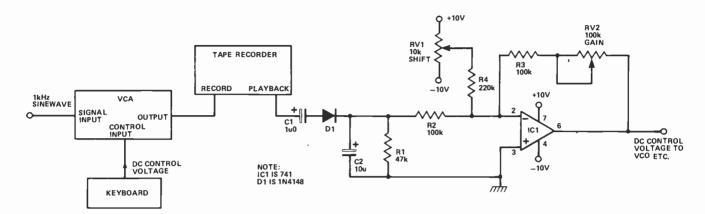
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TECH TIPS SPECIAL



Magnetic Tape Sequencer

P. Hill, Chelmsford

The following circuit enables a synthesiser sequencer to be constructed using a tape recorder as the storage device for the control signals. A1 kHz sinewave

is amplitude-modulated by the DC control voltage from a keyboard, for example, using a voltage-controlled amplifier. The output of the VCA is connected to a tape recorder and recorded.

On playback, C1 couples the AM signal to an AM demodulator comprising D1, R1 and C2. The demodulated signal is a replica of the original control

voltage and is passed to the buffer built around IC1, an op-amp wired in the inverting mode. The output is fed to the control input of a VCO, etc.

RV1 applies a DC shift to the output and can be used to vary the frequency range of the VCO. RV2 varies the gain and can be used to adjust the frequency/ voltage law of the sequencer.

One IC Dice

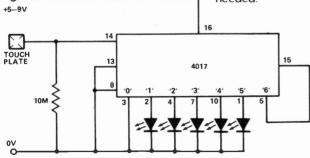
P. Heap, Cambridge

This must surely be the simplest possible electronic dice; it uses one IC, one resistor and six LEDs. Operation of the circuit is as follows. Normally the resistor keeps the input to the 4017 low and one LED is illuminated. When the plate is touched, mains hum is injected into the circuit by the finger. These input pulses are counted by the 4017, making the LEDs flash too quickly to see the number displayed. When the finger is removed

the 4017 stops counting and a steady display is obtained.

This principle has other applications: it could be used to clock a binary counter or applied to a gate, and used to provide a general-purpose 50 Hz clock. If the resistor is removed and the touch plate made large enough, sufficient hum is picked up to cause the circuit to count on its own.

The outputs of the 4017 could easily be decoded, using a diode matrix, to produce a proper dice display. However, this increases circuit complexity and extra driver transistors would probably be needed.



Triton Tape Modification

D.E. Buchan, Bristol

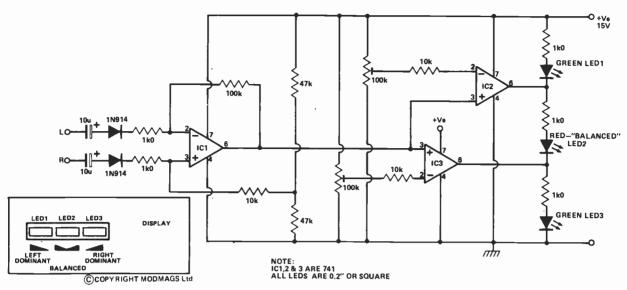
have found that the major problem with tape load on the Triton is that the filter can exhibit different phase responses with time, so that a program loaded one day would need several attempts the next day to get a good load. After the modification the load rate was 100%. The modification involves changing the filter from a narrow band-pass filter, centred around 400 Hz, to a band-pass amplifier with a bandwidth of about 3 kHz from 400 Hz to 3.4 kHz (the telephone system bandwidth).

The changes required are as follows: C28 and C31 are changed to 47n ceramic, C29 is removed from the board, and a 47n ceramic capacitor is connected between the collector of Q3 and +5 V. R39 (4k7) may have to be changed to reduce sensitivity (eg 12k).

Tech-Tips is an ideas forum and is not aimed at the beginner. We regret we cannot answer queries on these items.

ETI is prepared to consider circuits or ideas submitted by readers for this page. All items used will be paid for. Drawings should be as clear as possible and the text should preferably be typed. Circuits must not be subject to copyright. Items for consideration should be sent to ETI TECH-TIPS,

Electronics Today International, 145 Charing Cross Road, London WC2H OEE.



Stereo Balance Meter

G. Durant, Selby

Balance on a stereo amplifier is usually set by ear, but this of course can be very difficult to judge. If an amplifier has a balance meter at all, it is usually of the centre-zero moving coil type — bulky, old-fashioned looking and expensive. This circuit is designed to overcome all of these problems.

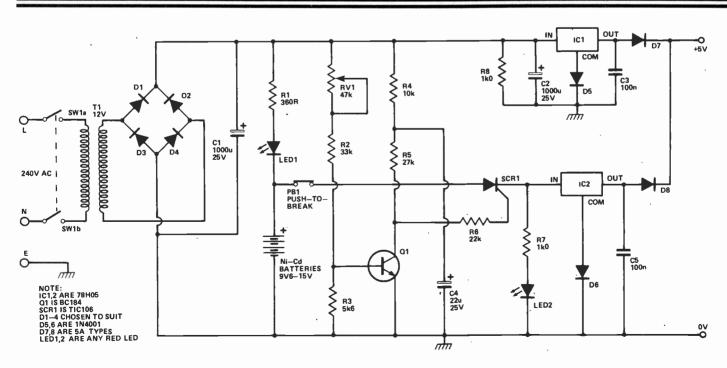
The outputs from each channel are fed to the two inputs of IC1, this being connected as a differential amplifer. If the left and right channels are of equal levels, the output of IC1 will have its output at about halfway

between the supply rails. If the left channel gets above the level of the right channel, the output of IC1 will approach the 0 V rail. If the right channel is loudest, the output becomes positive.

IC2 and 3 are also differential amplifiers, but in this case they are driven by the output of IC1. LEDs form a display at the outputs of the two ICs. Pin 2 of ICs 2 and 3 each go to a preset across the supply. In practice, the preset in conjunction with IC2 is set to hold pin 2 slightly above 0 V and the preset connected to IC3 is set to hold pin 2 just below supply voltage. These settings, however, must be set by trial and error so that the circuit works accurately.

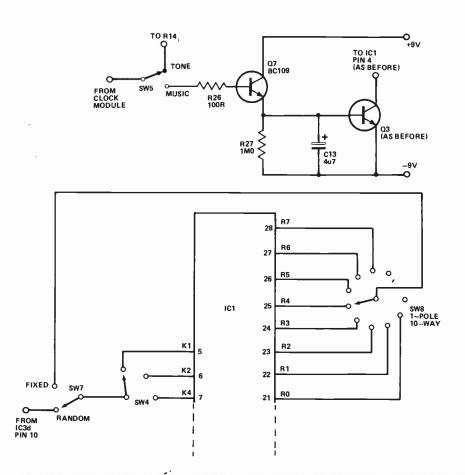
The output of IC1 is connected to the non-inverting inputs of IC2 and 3. If the output of IC1 approaches the supply rail, the outputs of ICs 2 and 3 will also go high, thus illuminating LED 3. This would happen if the right channel were dominating. If the left channel were dominant, the outputs of ICs 2 and 3 would be low, thus illuminating LED 1. If the two channels were equal in amplitude, the outputs of ICs 2 and 3 would be high and low respectively, lighting up LED 2.

The circuit can easily be added on to a ready constructed unit without using up large amounts of panel space, or used as a add-on unit for a hi-fi system. The unit draws about 20 mA, so battery operation is practical.



Alarm Clock Modifications

M. Rogers, Swansea



Musical Alarm Clock which make it even more versatile. The first enables an ordinary LCD alarm watch module to be used in the design, allowing the constructor to choose exactly which functions he requires, depending on the watch bought (eg day/date, snooze alarm, chronograph etc). The module is easily removed from the case once the back has been taken off — fine wire is soldered to the circuit board and switch plates to connect the module to the remainder of the circuit.

The first diagram shows the changes necessary to allow the musical chip to be continuously enabled by the pulsed alarm tone from the watch module. The alarm output is taken from the most positive of the small springs on the module when in the alarm condition. C13 may need to be increased if a watch unit with a long on-off bleep time is used. R12 and R13 in the original circuit are not needed and Q3 can be left on the PCB. Some constructors may feel that the decrease in display size is easily offset by the approximate halving of the cost of the clock.

The second modification allows you to select which tune is played by the chip, either in alarm or test mode. This is useful if you want a favourite tune played, or if you simply want to hear all the available tunes without waiting for them to be played randomly. There is enough room on the PCB for the pads for R_0 to R_7 ; if the board has already been etched then flying leads can be taken from the IC socket pins directly.

Computer PSU Back-up

E. Williams, Helston

One of the problems of running a microcomputer without auto-battery back-up is that, if the mains fails (usually after entering a long program), you lose the lot and have to re-enter it from the start. It only takes the 5 V supply to disappear for a few microseconds for the damage to be done.

After being caught once (and once was enough), I designed the supply shown here.

At switch-on the full-wave rectifier (D1-4) supplies regulator IC1, which gives 5 V after steering diode D7. D5 raises the output of IC1 to compensate for the drop across D7 (as does D6 for D8).

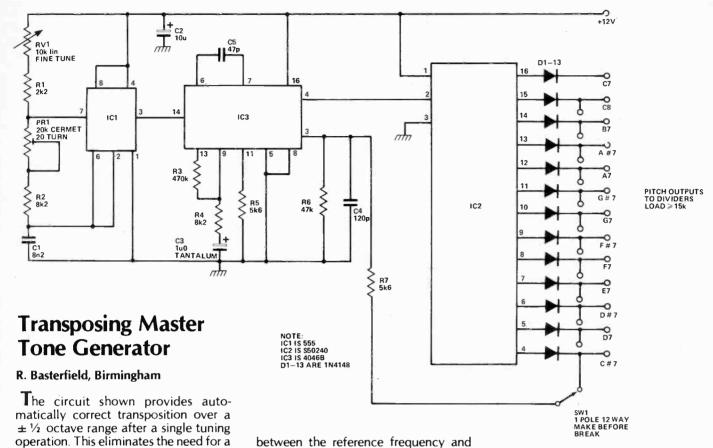
The collector of Q1 is normally held low, initially by C4 and R5 while C1 and C2 are charging to the full supply poten-

tial, and then by base bias network RV1, R2, R3. The low at Q1 collector holds SCR1 off via R6. The Ni-Cd batteries are trickle-charged by R1-LED1; LED1 lights up to indicate that the batteries are charging and becomes reverse biased when the unit is switched off to prevent battery discharge.

If the mains fails or is switched off, either by accident or on purpose, the current flowing into Q1 base via RV1 and R2 decreases to zero. Long before the supply to IC1 has become too low to maintain the 5 V rail, Q1 will have turned off and fired SCR1. This allows the batteries to supply IC2, a second regulator which takes over from IC1 before the microcomputer supply is interrupted. LED2 is turned on via R7 to give an indication that the batteries are being discharged.

If the mains should now come back on, LED1 will again light to indicate mains presence, but LED2 stays on until the reset button is pressed to switch off SCR1. When you wish to switch the unit off, first switch off the mains and then press the reset button to turn off SCR1. In this condition there is no drain from the batteries so the unit can be left for weeks without detrimental effect. Four amphour Ni-Cds were used to give a 9V6 supply. Two units have been built and both have proved 100% reliable.

If a supply of less than 1 A is required, 7805 regulators can be used and D7, D8 can be replaced by 1 A devices. The transformer and rectifier diodes should be rated in accordance with the current required. RV1 should be set to middle position and then adjusted for best operation. If battery voltages greater than 9V6 are used, R7 must be increased in value and R1 decreased. The transformer, bridge rectifier, IC1 and C1-3 may be available in your original supply, in which case it will only be necessary to modify it as shown in the diagram.



between the reference frequency and the output pitch selected by SW1. Any output selected by SW1 is always forced to produce the same frequency as the reference, so shifting the whole top octave. C4 filters RFI from the leads to SW1, which were 5" long in the prototype.

To set up, select G#(pin 11) with SW1. Set RV1 central. Adjust PR1 to give A = 440 Hz or IC1 output pin 3 to give a 300.7 us period.

The \$50240 is available from Tandy Corporation.

Micropower LED Flasher

D. Stewart, Wick

expensive presets.

This circuit will brightly flash an LED, yet has a supply current of only 150 uA. In a normal 555 astable, the timing capacitor is discharged straight to ground. Here, the charge is made use of by discharging it through the LED. A suggested use is for an on-off indicator in a battery-powered circuit.

trained musical ear to set a bank of often

of 3.32 kHz. IC2 is a top octave syn-

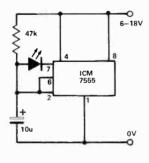
thesiser and IC3 is a CMOS phase-locked

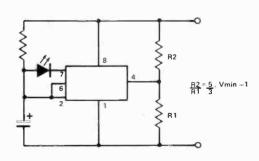
loop, producing a high frequency on pin

4 in proportion to the phase difference

IC1 generates a reference frequency

With a slight modification the circuit can be used as a good battery indicator. A potential divider is connected to pin 4 (reset) from the supply rail of the circuit whose battery is being monitored, so that when the supply drops below a predetermined voltage, then the voltage on pin 4 drops below 0V7. Thus the LED will only flash if the supply is higher than the predetermined voltage. Keep the value of the resistors high to reduce current consumption (eg 1M0 for R1).





FEATURE: Tech Tip Special

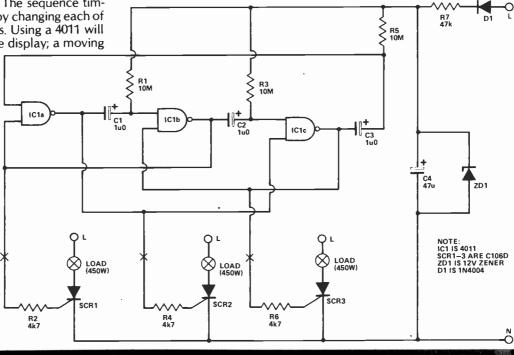
Cheap Light Sequencer

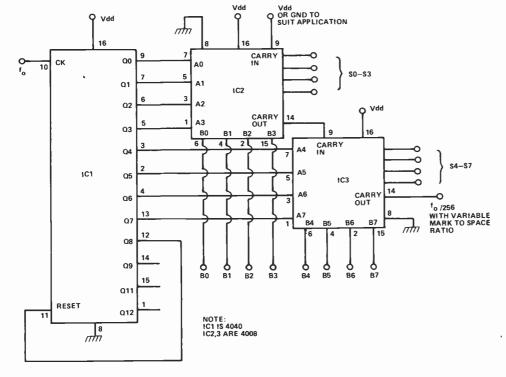
Kevin Kirk, Malta

This light sequencer is basically an extended monostable. The sequence timing can be changed by changing each of the timing capacitors. Using a 4011 will create a moving hole display; a moving

light display can be obtained by replacing the 4011 with a 4001 or putting an invertor at the points marked 'x'. If an EXNOR is placed here a single switch may

be used to change from one state to the other. With the values given the sweep rate will be very slow and is used for a slow sign changer.





Digital Mark/Space

G.C. Dean, Taunton

I his circuit provides a mark/space ratio at the C_{OUT} pin which depends on the binary value set up on B0 to B7. As Q0 to Q7 gradually increases in value, due to incoming clock pulses, $C_{OUT} = 0$ if $Q0...Q7 + B0...B7 \le 11111111$ and $C_{OUT} = 1$ if Q0...Q7 + B0...B7 > 11111111. The higher the value of B0...B7, the quicker C_{OUT} will become 1 after Q0...Q7 is automatically reset, and the higher the value of the mark/space ratio. The proportion of time that C_{OUT} is 1 is given by:-

$$(\frac{\text{Value of B0...B7}) + C_{IN} (= 0 \text{ or 1})}{256}$$

Note that for C_{OUT} to be permanently 0 C_{IN} must be 0 (and B0...B7 = 000000000) and that for C_{OUT} to be permanently 1 C_{IN} must be 1 (and B0...B7 = 111111111).

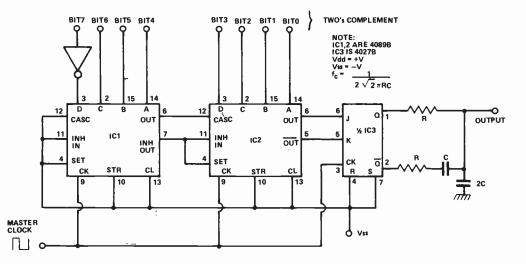
The circuit could have its clock input connected to a microprocessor clock, B0...B7 connected to the data bus and C_{OUT} to a moving coil meter or a red/green LED (RS 587-080). Then the meter reading will be proportional to, or the colour of the LED will depend on, the value of B0...B7.

D-to-A (By Stealth)

S.R. Gillbard, Liskeard

f T his D-to-A converter uses 4089Bs, which are four-bit binary rate multipliers. Each rate-select input controls an internally generated pulse train, with frequencies of 1/2, 1/4, 1/8 and 1/16 the clock frequency. All pulses coincide with a low state of the clock signal. The rate output is formed by interleaving the selected pulse trains. The output of IC1 consists of N_H pulses per 16 clock cycles, where N_H is the high-order nibble of the eight-bit word (hence $N_H \leq 15$). The sixteenth (blank) clock cycle allows space for the low-order section, IC2, to insert up to 15 pulses per frame (256 clock cycles) as determined by N_L , the low-order nibble. Due to propagation delays the pulses at the IC2 'rate' output, when present, will persist for a short while after the clock's rising edge. IC3 makes use of this feature to extend each pulse so that it completely fills one clock cycle and hence merges adjacent cycles at like polarity. More important, IC3 re-synchronises the transitions between high and low levels, according each 'pulse interval' the same significance.

The resulting waveform may be considered to be a distributed pulse-width modulation in which the maximum continuous duration at the 'wrong' polarity is equal to one clock period, the error thus being dispersed throughout the conversion cycle. This should be contrasted with a strict digital implementa-



tion of PWM in which the signal would be high for number of adjacent clock cycles and low for the remainder of each frame. It will be noticed that, while both types have the same average level and would require the same bandwidth for accurate transmission, the standard form has much larger low-frequency components. This is most strikingly illustrated by the conversion of a digital zero. PWM produces a 50% square wave with a frequency equal to the frame rate; the distributive technique generates a square wave at 128 times frame rate! While other conversions provide increasingly less marked contrasts, it is only at the extreme limits of ± 127 that the two systems return identical results.

This feature of the distributive technique permits the use of quite simple averaging filters with relatively high cut-off trequencies and hence fast rise times. The four-component network shown, when used with a centre frequency equal to half the frame rate, will suppress ripple to more than 80 dB below peak reconstructed sine wave.

The digital input is completely asynchronous, therefore any data feed below frame rate is acceptable. Faster data transfer will result in a slight loss of precision and low-level intermodulation between frame and feed rates. If only sevenbit conversion is required, the inverter should be driven in parallel with the highorder 'C' line, producing a half-range output.

Synthesiser Interface For Transcendent DPX

D. Pallant, London

This circuit was designed to interface a voltage controlled synthesiser such as the Transcendent 2000 to the Transcendent DPX.

The six multiplex lines, AO1 to AO6, are converted into a staircase waveform by the D-to-A converter, IC1, and are also NANDed by IC2 to give a 'keyboard scan finished' signal. The converter output is fed into two sample and hold circuits which sample the waveform (IC5,6,7) and output it to the buffer and level converter IC8. When SW1 is open the first sample and hold will remember the voltage on the D-to-A converter when the scan reaches the lowest note pressed on the keyboard. The logic in this case is performed by a latch and a NAND gate

which output a pulse when the first note is pressed but, due to the latch, ignore all subsequent notes until the latch is reset by the scan finished signal. The second sample and hold is triggered by the 'scan finished' signal from IC3a, and will therefore output the voltage on the first sample and hold at the end of each scan.

If $S\overline{W1}$ is closed the first sample and hold will remember the voltage of the last note pressed all the time. This is because it is then triggered solely by CIN, which goes low as the scan reaches each key pressed. At the end of the scan the second sample and hold will then output the voltage proportional to the highest note pressed. The buffer on this circuit is widely variable and could be used for most VCOs. The volts per octave scale is set by PR1 and the tuning by RV1, both of which should be set up for the VCO to be used with the circuit. A suitable trigger pulse for the Transcendent 2000 can be obtained by inverting the key pressed (KP) signal which can be obtained from the output of IC7 (pin 9) on the dynamics board.

The circuit can be easily fastened inside the right-hand end cheek of the DPX above the dynamics board, and two ¼-inch jack sockets put on the back panel for the voltage control and trigger outputs. The logic signals required are obtained off the dynamics board as follows:

AO0 IC17, pin 8 AO1 SK1, pin 5 AO2 SK1, pin 4 AO3 SK1, pin 3 AO4 SK1, pin 2 AO5 SK1, pin 1 IC6, pin 4

In almost all cases there is a throughboard pin that can be used as a terminal point.

Bidirectional Audio Link

T.P. Hopkins, Stockport

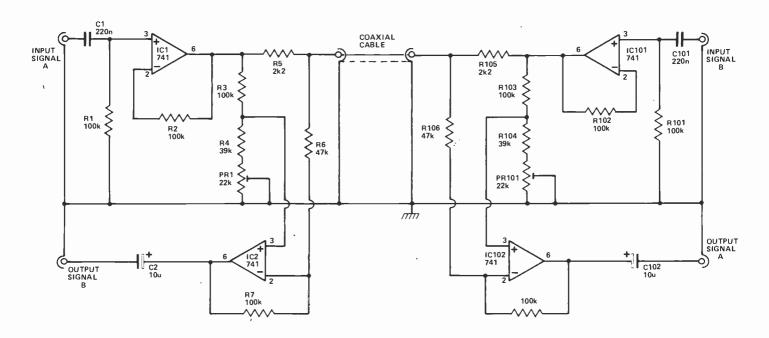
This simple circuit arrangement enables audio signals to be sent along a single piece of coaxial cable in both directions simultaneously. The circuit consists of two identical parts, one in each device connected. The input signals are buffered by IC1, IC101 and

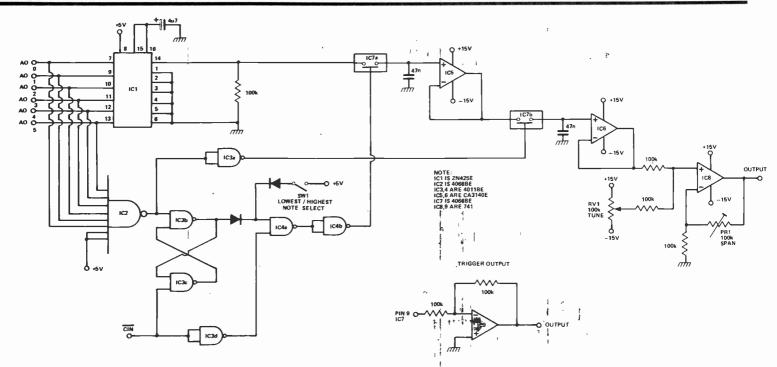
fed to the cable by resistors R5, R105. IC2, IC102 subtract the signals on the cable from the output of the buffer amplifier; the difference is the signal put onto the cable at the other end. The net result is that signals inserted at one end appear only at the other end.

The audio signals should be between 100 mV and 3 V (RMS). Potentiometers PR1, PR101 set the rejection of the unwanted signal; these should be of good quality, and preferably multiturn presets. A rejection of 50-55 dB can be

obtained.

The prototypes were used in an audio system where the control unit was remote from the signal source and the power amplifiers and speakers. Other possible uses include intercom and talkback systems. If this technique is tried at higher frequencies, resistors R5, R105 should be adjusted to match the characteristic impedance of the coaxial cable used. A similar system has been successfully used for digital signals.





ETI JUNE 1981

FEATURE: Tech Tip Special

Memory-mapped Sound Generator

W.S. Maggs, Bristol

Here is a circuit idea for using AY-3-8910 Programmable Sound Generator chips on an Apple bus.

The usual way of interfacing these chips to microprocessors (other than PIC 1650) is by using PIAs. However I wanted to remove the additional burden and wastefulness of programming PIAs by memory-mapping the PSG. This is done as follows; from the data sheet pins BC1, BC2 and BDIR control the chip operation (see Table 1).

It can be seen that if BC2 is taken to +5 V then only BC1 and BDIR control the chip. I tried various gate layouts but these failed. Finally I used a dual 4-line to 1-line data selector (74LS153), each side controlling BC1 and BDIR respectively.

To control the '153 I used the R/\overline{W} and A0 lines of the Apple on the A and B inputs. The inhibit input was selected by a signal from the 7442 so that the '153 was only selected if the '0' output of the 7442 was low (see Table 2).

The selection of the 7442 is an easy matter on an Apple bus because each slot has a line DEV which selects one of 16 addresses. To make programming easier the reset pin of the PSG is also memory-mapped. The OR gate on output '1' of the 7442 makes sure the reset is only selected by one address.

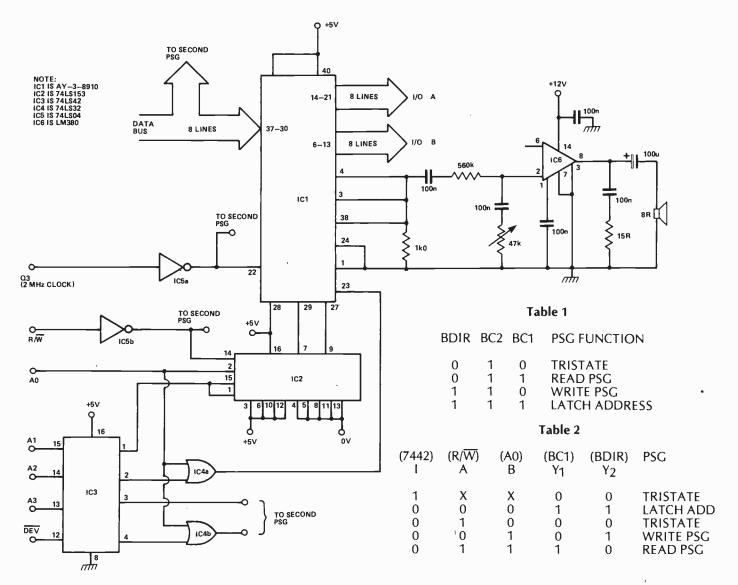
A simple program to test the chip is given. Since the PSG requires a register address first, then the data for the register, the register address (latch address) is mapped C0X0 (where X is the slot number + 8). The data is then read or written to C0X1 with the reset on C0X2. Resetting the PSG clears all registers. In the program the data is held at 2000 Hex, with the number of registers used in

2000, followed by the data and then the registers (hi → lo). To reset the PSG a dummy store is used.

Since the load requirements of the Apple bus are fixed, the R/W line is buffered to enable two PSGs to be used. The I/O ports of the PSG can either be used as extra PIAs or to address PROMs. The signal outputs of the three channels of the PSG drive a conventional audio amplifier (LM380).

Program for frequency 440 Hz

1000	8D C2 C0	STA \$ C0C2
	AC 00 20	LDY \$ 2000
	BE 04 20	LDX \$ 2004, Y
	B9 00 20	LDA \$ 2000, Y
	8E C0 C0	STA \$ COCO
	8D C1 C0	STA \$ COC1
	88	DEY
	D0 F1	BNE \$ 1006
	60	RTS
2000	04 OF 3E	00 FE 08 07 01 00



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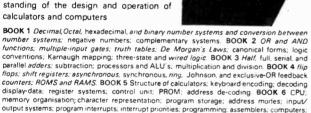
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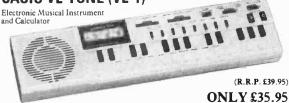
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ALIEN ATTACK

Hand-held games are becoming just as popular as the arcade versions (we can play in the comfort of our own office!) so we just had to publish our own version. Design and development by R. Eley.

nce upon a time you could only blast the hordes of little green aliens by taking a trip to your nearest pub or amusement arcade. But it wasn't long before you could indulge your violent tendencies in the privacy of your own home with TV game versions. Nowadays it's possible to avoid withdrawal symptoms wherever you may be by purchasing a hand-held version — the invaders are even turning up in calculators. Now ETI presents a simple-to-build hand-held game that, while lacking the refinements of commercial machines (such as custom-designed little 'alien' LEDs), is still a lot of fun to play with and offers a full range of sound effects.

The 'field of battle' and the score display both take the form of a line of LEDs. When the game is switched on, 'aliens' begin to drop towards you, their passage being shown by the LEDs in the display lighting one after another. When the tenth and final LED is lit, you have to fire your laser at the alien by pushing the 'fire' button. If you're successful, the score display is increased by one and another alien launches his attack. For simplicity and low cost, a simple binary counter is used to register the score.

The catch is that as you destroy the aliens, the speed at which they fall increases quite rapidly. The game has a built-in time limit of about 25-30 s, and the object is to achieve the highest score before the game ends. Your reactions have to be pretty accurate because firing the laser when the ninth LED is lit will zero your score.

Four voltage-controlled oscillators are provided, giving the familiar tromp-tromp, laser fire, falling bomb and explosion noises. An on-off switch is provided for the sound so that battery life may be extended, if desired. The unit consumes approximately 15 mA with sound or 5 mA without.

Construction

The circuit is built on a single PCB but for reasons of space this is fairly cramped and several components are mounted vertically. Tantalum capacitors are also used instead of ordinary electrolytics because of their small size. Solder all the components in place as shown on the overlay, using a soldering iron with a fine bit and lots of 'due care and attention'; the PCB tracks are very fine. Take the usual precautions when handling the CMOS ICs.

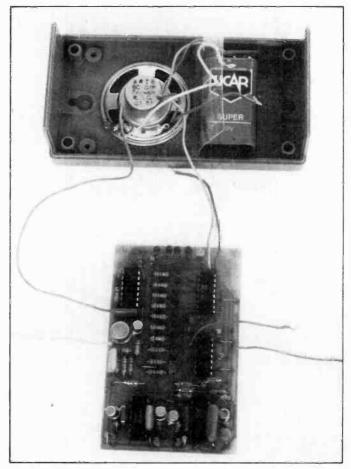
Note that R29 and C13 are not located on the PCB but are soldered into the loudspeaker lead — the photographs should make this clear.

A T-shaped hole is cut in the top of the case to reveal the LEDs and a piece of red plastic can be stuck over the aperture to improve the viewing contrast. Of course, you'll have to cut holes in it above the green LEDs or they'll disappear! The three switches are also mounted on the top of the case; the loudspeaker is fixed to the bottom after drilling a few holes to let the sound out. Thin plastic strips are glued to the sides of the case to support the PCB the correct distance from the cutout.

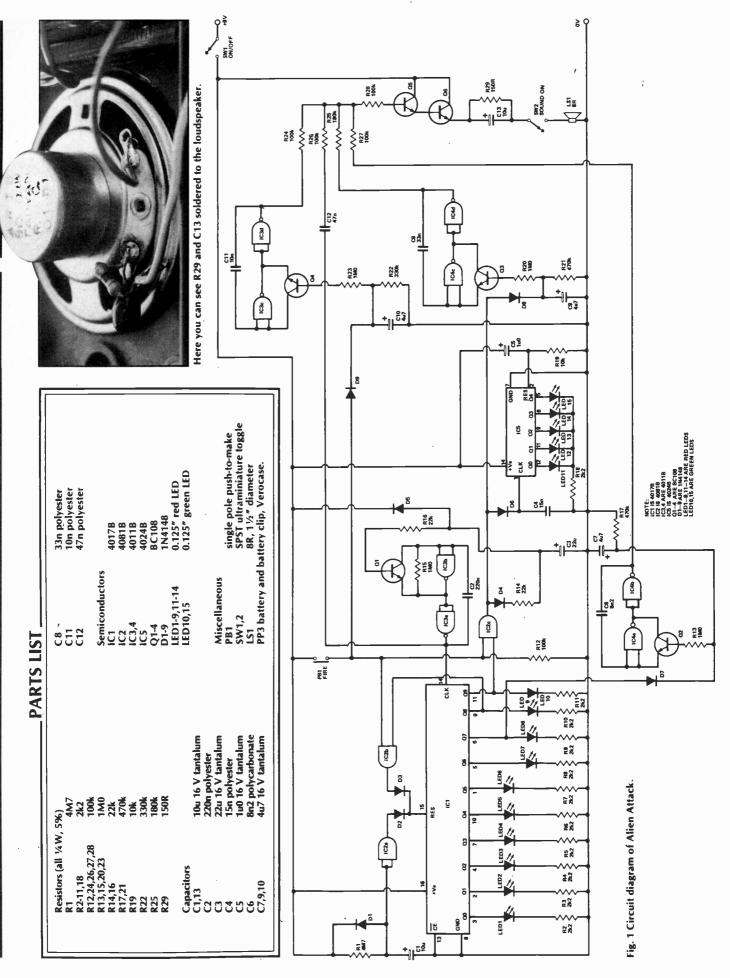


Now the interwiring can be completed and the case screwed together.

This completes the construction of the project; now you can be the envy of your fellow commuters and annoy total strangers in your efforts to beat your last score.



The completed board, wired up to the lower half of the case. The remaining wires go to the switches on the front panel.



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A. A.

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HOW IT WORKS

which lights LEDs 1 to 10 in turn; the binary counter ICS, which provides the scoring for the game; four voltage-controlled oscillators VCOs) which provide the sound effects; and the mixer-amplifier, circuit falls into four basic sections - the decade counter IC1 which drives a small loudspeaker.

difference. Instead of using a fixed resistor with a capacitor to deter-The VCOs use the common CMOS oscillator circuit, but with a mine the frequency, a transistor replaces the resistor and functions as a

Taking the VCO formed by IC4a and IC4b as an example, it can be seen that with no connection to the base of Q2, the collector-emitter resistance will be very high, preventing C6 from charging and thus disabling the oscillator. However, if a voltage is applied to Q2's base the collector-emitter resistance will fall in proportion to the applied voltage. Thus the time taken for C6 to charge will be proportional to this voltage, and so will the frequency of operation of the oscillator. variable resistor

If a capacitor and resistor are connected from the base of the transistor to ground, then fully charging the capacitor will give the highest oscillator frequency. As the capacitor discharges via the resistor, the frequency will fall until the circuit again stops oscillating.

In the case of the IC3a-IC3b VCO, the lowest frequency is determined by R15.

pulse to IC5, thus extinguishing LEDs 11 to 15. C1 will also start to When the circuit is switched on, C5 provides a power-on reset charge via R1 and when the voltage on C1 eventually reaches the threshold of gate IC2a counter IC1 will be held reset, thus ending the game. With the values shown, this should take approximately 25 s.

The IC3a-IC3b VCO will clock IC1, lighting LEDs 1 to 10 in turn. When LED8 lights a pulse will be fed to the VCO formed by IC4a and IC4b, giving a falling frequency.

enabled. Pressing PB1 when LED10 is lit will result in IC2c enabling the If the 'fire' button (PB1) is pressed, then the IC3c-IC3d VCO will be determined by R14. Thus the VCO driving IC1 will increase in frequency. IC5 will also be clocked, adding one to the score. C4 debounces the VCO formed by IC4c and IC4d, and also charging C3 by an amount clock input.

If PB1 is pressed when LED9 is lit, a reset pulse is sent to IC1 by

IC2b, thus preventing cheating.

by R29 and C13. This filter prevents excessive DC from reaching the The four oscillator outputs are mixed by R2427 and C12. Q5 and Q6 act as an amplifier, driving an 8R speaker, through the filter formed speaker, as would happen if one of the VCO outputs stayed high.

results in the aliens making several abortive attacks. D1 and D5 ensure Because IC2a has no hysteresis applied to the input, as the voltage on C1 reaches the gate's threshold the output will oscillate, which that capacitors C1 and C3 are discharged at the end of each game when the circuit is switched off. This ensures that the game length and starting speed of the aliens are the same for each game.

BUYLINES

No problems here, as everything used is pretty common. The case used was a small one from Vero, reference no. 202-21029J, and the red plastic can either be a special LED display filter or any piece of cellophane you can lay your hands on (sweet wrappers?).

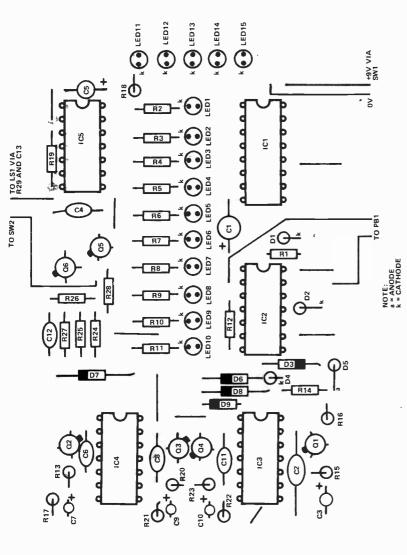
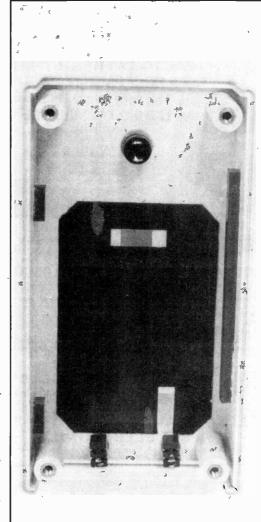


Fig. 2 Component overlay.



This photograph shows the red filter and plastic mounting strips glued into the top of the case.

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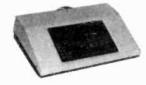
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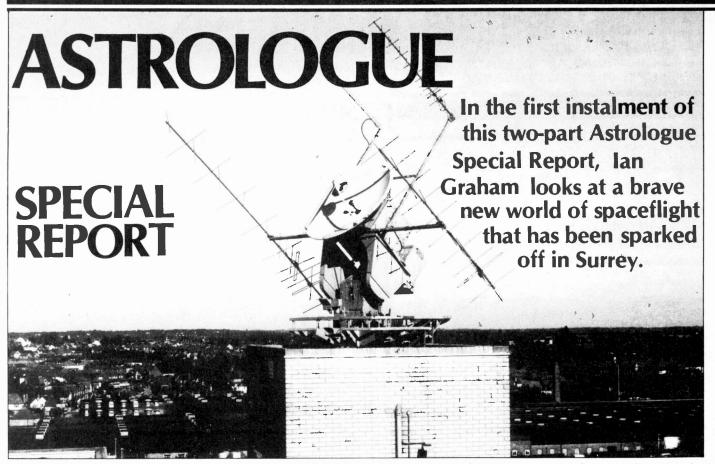
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ou read it first in Astrologue. In the December 1980 edition I reported on Britain's first amateur satellite — UOSAT (University of Surrey Satellite). Now it's headline news as the launch day draws near. I spent a very enjoyable day as a guest of the project team at the University. Their work is so impressive and, in my opinion, has such important implications for Britain's future involvement in space that I am devoting two editions of Astrologue to UOSAT. This month there's a general view of the satellite and its objectives, and next month a rundown of the technical data.

New Look

The team, headed by Dr Martin Sweeting, has taken a refreshingly new look at the business of designing a spacecraft. They haven't automatically followed the conventional solutions to problems, largely because of cost. The last satellite of similar size was UK6, which was launched in 1978 at a cost of about £9 million excluding the launcher. The Surrey team is working on a budget of £100,000.

The honeycomb structure that forms part of the spacecraft surface would normally have cost £15,000, but the team found a firm which produces the same material for racing cars and which would do the job for £300.

The team has explored every avenue in an attempt to cut costs, but that doesn't mean that they are turning out a Heath Robinson stringbag affair of a spacecraft. Before NASA will agree to launch the craft they must be satisfied that it passes all the vibrational and thermal tests, structural specifications, etc demanded of every satellite.

Genesis

How did they get into the satellite business? Five or six years ago a small tracking station was set up to track OSCAR 7 (an American amateur radio satellite) and meteorological satellites. It was a natural progression from the work on tracking, amateur radio and ground station operation to actually building their

own spacecraft. The driving force in the early days was the University's Electronic and Amateur Radio Society (EARS — ho, ho), which made the initial contact with AMSAT (Radio Amateur Satellite Corporation) and set up the first ground station.

The present ground station has cost about £10,000 over 10 years. The impressive aerial array is mounted on a Bofors gun platform rescued from an Admiralty scrapyard in Portsmouth. The whole assembly, weighing between 3 and 5 tons, can be pointed at any object in the sky and tracks satellites either manually or by computer, using a control system designed by undergraduates. The precarious, lofty perch is necessary so that signals can be received without ground interference from street lighting, traffic, and so on.

Project UOSAT

The UOSAT programme has three broad interests:

- 1. Examining low-cost techniques
- 2. Educational
- 3. Research.

The determination to use low-cost techniques where possible has extended beyond the practical (or practically impossible) business of designing and building the spacecraft. The aim from the beginning has been to make the satellite information available to as many people as possible, as cheaply as possible, with emphasis on the involvement of schools and colleges.

Data will, therefore, be available on a number of levels. In its simplest form, you will be able to receive spacecraft telemetry on an ordinary VHF communications receiver. This is made possible by the provision of an on-board speech synthesiser linked to selected subsystems. The spacecraft will 'speak' its status to you. If you have the necessary equipment, you can also receive the information in Morse code. Data will also be available in high-speed form for serious enthusiasts and professionals.

FEATURE

Left: our lead photograph shows the EARS aerial array linked to the command centre, which will be the primary command station for UOSAT. Right: the UOSAT logo. Far right: the team built their own clean room in which to construct the spacecraft. Project Manager, Dr Martin Sweeting, is on the right.



Telly Weather

More exciting to the likes of you and I, however, is the meteorological experiment. The spacecraft will photograph an area the size of Scotland as it orbits. Normally you would need several thousands of pounds worth of decoder and display/printing gear to get a weather picture out at the other end. However, the UOSAT team hope to be able to produce a handbook and kit (being designed by undergraduates) to allow weather pictures to be displayed on your living room telly. The kit is expected to cost no more than £150. Watch Astrologue for a review.

The satellite is expected to be launched in September but the team has been warned that the launch date might be brought forward to July, leaving them very little time. A series of tests will be carried out in June at British Aerospace, Stevenage. Then two days of final tests must be carried out at the Goddard Space Centre in Maryland before the launch from Vandenberg Air Base in California.

Piggy Back

UOSAT is being given a 'piggy back' launch; the main payload of the Delta 2310 launcher will be the NASA Solar Mesosphere Explorer. UOSAT will be a secondary payload, fitted beneath and to one side of the main payload. It will orbit at a height of 530 km with a period of 95 minutes and should stay up for four or five years before it re-enters the atmosphere in a fireball.

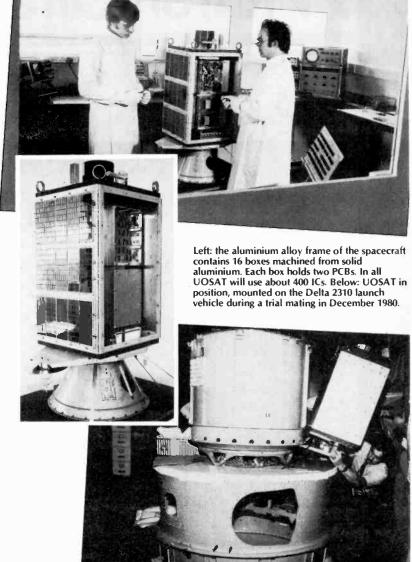
Milestone

Up to now we have assumed that building and launching a satellite is a multi-million pound/dollar operation. The UOSAT project team has proven that the cost can be reduced by at least one order of magnitude without compromising quality. The team has actively sought out sponsorship (money or test facilities) from the beginning of the programme. It is doubtful whether sponsorship would be as forthcoming for further projects of this nature, but once the principle of low-cost spaceflight is proven, it is undoubtedly a concept that could be turned into a commercial enterprise. If cost-to-orbit can be reduced from £10 million to less than £1 million, that bodes well for future British involvement in an active space programme, in which more and more people can participate.

This single project at Surrey University looks like revolutionising our approach to spaceflight. It's an important milestone in the 'space age'. UOSAT demonstrates that a small team of enthusiasts working on a very slim budget can make a valuable contribution to the use of space.

A final word — if you want any information on UOSAT, please don't try to contact the project team. They are working to a very tight schedule indeed to ready the spacecraft for its possible July launch.

Next month — a structural and technical run-down of UOSAT and its experiments.



SHORTS

British Aerospace Dynamics Group and Plessey are to collaborate on defence communications projects. In the past, the Dynamics Group has participated in over 45 communications satellite programmes (10 as prime contractor). Plessey designed and manufactured Skynet — the first operational ship-borne satellite communications terminal — and is involved in satellite earth station manufacture.

British Aerospace will supply the spacecraft and provide overall leadership of the programme. Plessey will supply earth station equipment for monitoring and controlling the spacecraft and interfacing into terrestrial networks.

The Vatican has finally caught up with the twentieth century. At the request of Pope John Paul II, it is to review Galileo's heresy conviction of 1633. The conviction arose because Galileo contradicted contemporary beliefs (scientific and theological) when he proved that the Earth was not the centre of the Universe — it orbited the Sun like any other planet.

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VOLTAGE-CONTROLLED AUDIO

DC control of hi-fi — Keith Brindley looks at some ways and means of achieving it, using Mullard's range of voltage-controlled ICs.

ack in the old days, an amplifier was an amplifier! It would have simple controls; volume, bass, treble, balance and a mode switch. The most expensive amplifier was the one which did its job (ie amplify) with the least amount of noise and distortion from linearity.

All of a sudden, manufacturers (particularly those on a faraway eastern island) got the knack of bettering their specs and the market became flooded with reasonable quality, reasonably priced systems. "Ah so, what next?" they thought, and the answer was gimmicks — little tricks which they could perform on an amplifier to make yours ever so slightly better than Mr Jones' next door, and which makes Mr Jones want to buy another one slightly better than yours.

In The Distant Future?

Two of the latest gimmicks are remote control and touch control (not just of amplifiers, but of TVs, videos and, presumably in the future, complete house electrical equipment). Now, most gimmicks — scratch filters, rumble filters, loudness controls and so on — don't add anything to the actual quality of the device; they merely colour the sound to suit the individual listener. To the hi-fi freak they are little more than useless!

However, remote or touch control can be highly desirable to the audiophile. You see, to control things at a distance or at a touch we need to be able to make all the adjustments with a mere change of voltage. In this way all the mechanical switches and potentiometers become obsolete. Even if mechanical controls are used they do not carry the AC audio signal, but only a DC control voltage. This means that once the signal is on the circuit board it stays there until the output. There are no signal-carrying leads to and from the pots, switches, or other controls and hence there will be less interference pickup, less interchannel crosstalk, better frequency response and fewer switching clicks and crackles — which means a dramatic improvement in amplifier quality.

Control Yourself

Recently, Mullard introduced a range of integrated circuits intended for use on DC-controllable, audio-frequency amplifiers whereby all the (usually) mechanical functions of

preamplifiers are controlled by DC voltages on particular IC pins. The two ICs of interest here, the TDA 1028 and TDA 1029, are electronic switches which fulfil the functions of mode switches, filter switches, mute switches and so on. A discussion of further ICs in the range (which control volume and tone) is planned for future presentation.

The TDA 1028 contains two double-pole, double-throw switches and the TDA 1029 one double-pole, four-way switch. Figures 1 and 2 show the simplified internal block diagrams of the ICs. If you bear in mind that these switches, although primarily intended for small-signal AC work, will also accurately switch analogue DC voltages, you will see that they are extremely versatile. The applications given later are all audio-frequency AC designs, but obvious DC suggestions lie in test or measuring equipment.

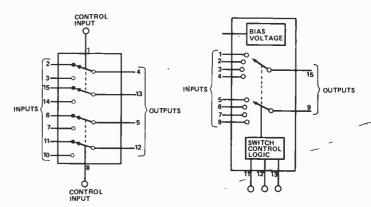


Fig. 1 Block diagram of the TDA 1028, a dual double-pole, double-throw switch.

Fig. 2 Block diagram of the TDA 1029, a double-pole, four-way switch.

A Look Inside

Figure 3 shows a more elaborate diagram of one of the internal signal paths. Basically, the input stage operates with a peak-to-peak input signal of something less than the supply voltage (ie typically between 3 V and 19 V). If a signal outside

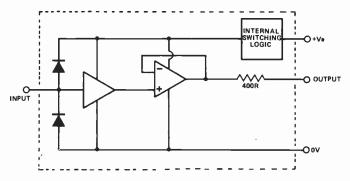


Fig. 3 Block diagram of a single internal signal path.

the supply range is applied, the diodes limit the signal, thus protecting the device.

A high impedance buffer follows the input stage and the output is connected via a 400R internal resistor which gives protection in the case of a direct short circuit. Overall gain of the signal path is close to unity and depends on the output load. For example, with a load impedance of 4k7 the gain is -1.5 dB(x0.84). Switching between one input and another is done by the internal logic, performing the simple function of connecting or disconnecting the power to the signal path in question.

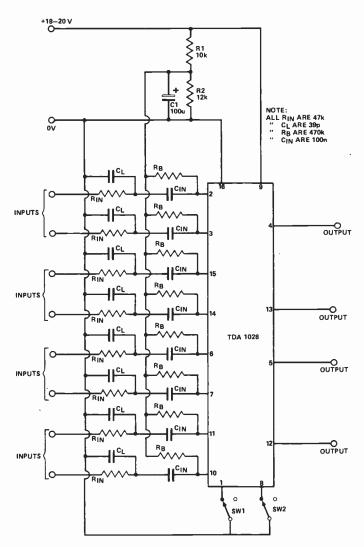


Fig. 4 Minimum component circuit for the TDA 1028.

Complications

Naturally, life is not so easy as the block diagrams suggest and certain peripheral components are needed to get the devices up and running.

(i) If the input voltage range (ie about 3 V to 19 V) is exceeded then the input current must be limited by an external resistor. The value of this resistor should be calculated so that the average input current does not exceed 20 mA. In the case of the ICs' use within a preamplifier no limiting resistors are strictly necessary.

(ii) If the switches are intended for AC work then DC block-

ing capacitors should be used.

(iii) A 'floating' input might cause switching noise at the output, due to rapid DC variations. To prevent this all inputs should be biased, via a resistor, to a point midway between the input voltage limits — about 11 V.

Workable circuits, with the above points considered, are found in Figs. 4 and 5. Bias for the TDA 1029 is supplied from an internal reference. This reference is unstabilised and dependent on the supply voltage. Hum and other interference will therefore affect it so a filter capacitor (C1 in Fig. 5) is required. If the TDA 1028 is used in the same circuit as the TDA 1029 then all input biases can be taken from this internal source. Alternatively, or in a separate circuit, a simple voltage divider and filter capacitor will do the job (R1,R2,C1 in Fig. 4). Bias is supplied to all inputs via 470k resistors (R_B) and all input capacitors (C_{IN}) are 100n polycarbonates. For experimental work only, all input resistors are specified as 47k, and low-pass capacitors (C_{IN}) can be used to eliminate RF interference in the input signal leads if required. These two circuits can be built for test purposes on the

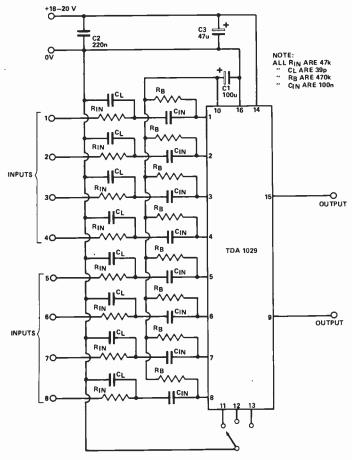


Fig. 5 The corresponding basic circuit for the TDA 1029.

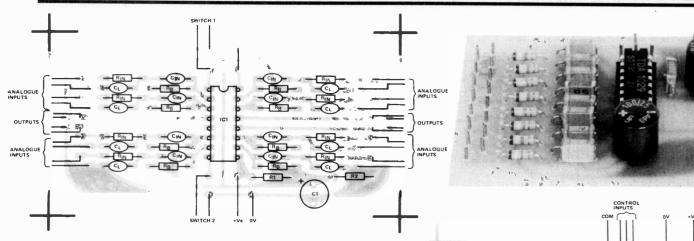


Fig. 6 Suggested layout for the experimental circuit of Fig. 4. The PCBs are on page 96.

PCBs shown in the overlays in Figs. 6 and 7. Adjust the input resistors as required (47k is a suitable starting value), and for DC applications use links instead of input capacitors. Incidentally, the boards are also suitable as bases if you intend to breadboard some of the following applications.

The final things to look at before using the two chips are the control inputs: they are internally connected to a HIGH state and the simplest way to ensure the LOW state of an input is to connect it directly to 0 V. Voltage control of the inputs requires <2V1 for a low state and >3V3 for a HIGH state—the inputs are therefore compatible with CMOS or TTL logic, and remote. touch, or even computer control of the signal paths is easy.

Table 1 gives a summary of the control inputs and consequent pin interconnections for the TDA 1028. Similarly, Table 2

gives the same for the TI priority over those of pilover pin 13, but not pin	ns 12 and 13.	Pin 12 cont	rol has priority Fig. 7 Overlay for the Fig. 5 circuit.
Connected pins	Control voltages		-times - times - t
Connected pins	Pin 1	Pin 8	
2 — 4, 15 — 13	Н	_	The state of the s
3 — 4,:14 — 13	L	_	
7 - 5, 10 - 12	 -	H	
6 — 5, 11 — 12	_	L	

ANALOGUE

ANALOGU

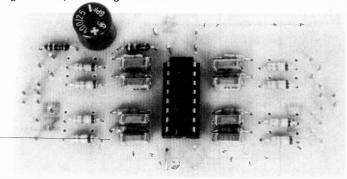
Table 1 TDA 1028 interconnections and control levels.

Applications

Perhaps the simplest use of the TDA 1029 is as a four-input stereo signal-source switch connecting either a stereo tuner, phono, tape deck or auxiliary input to an amplifier. Although it all sounds rather complex, Fig. 8 shows that it is not. The tuner and auxiliary input signals are fed directly, via input capacitors, into the IC, however, the tape output has a relatively high impedance source and connecting leads are therefore quite susceptible to RF interference. A suitable input network (eg R1 and C1) connected as a low-pass filter eliminates the RFI. The output of the circuit is fed back to the tape deck via coupling capacitors and 820k resistors.

The pick-up input needs RIAA equalisation and amplification; this stage is shown as a block in the circuit. Suitable circuits for pick-up stages are common, and no design for such is offered here.

It is possible to cascade the electronically switched signal paths of these two ICs, either within the same device (eg from



The TDA 1028 PCB. Note that the low-pass capacitors are optional and have been left off our board.

one half of a TDA 1028 to the other half) or from one IC to another. In either case separate signal paths are best connected via capacitors to keep switching clicks as small as possible. Figure 9 shows a typical example of a four-input stereo signalsource switch (using a TDA 1029) cascading into a monitor and stereo/mono switch (formed by a TDA 1028). A monitor switch allows comparison of recorded/played-back signals to and from a three-headed tape recorder. Thus it needs to be after the main signal-source switch but before the power amplifier, as shown.

The final two circuits given as application suggestions are switchable active filter circuits. Because each signal path is basically a unity-gain, non-inverting amplifier it is relatively easy to connect into standard filter circuits (a third-order Butterworth design being chosen) allowing electronic control of the preamplifier's frequency response.

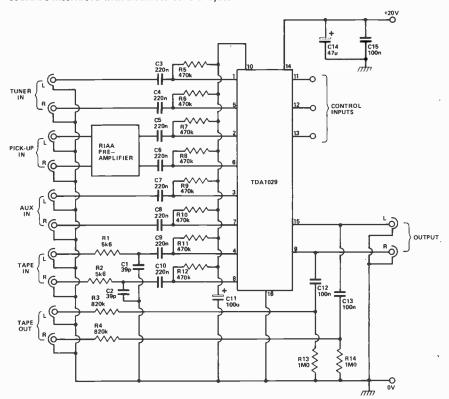
By using each switch section of a TDA 1028 in separate filter modes (Fig. 10) a stereo high-pass/low-pass (rumble/ scratch) filter can be built.

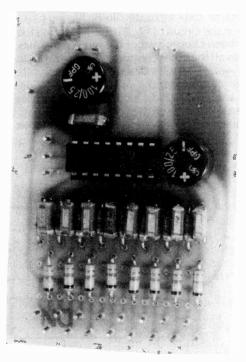
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SWITCH

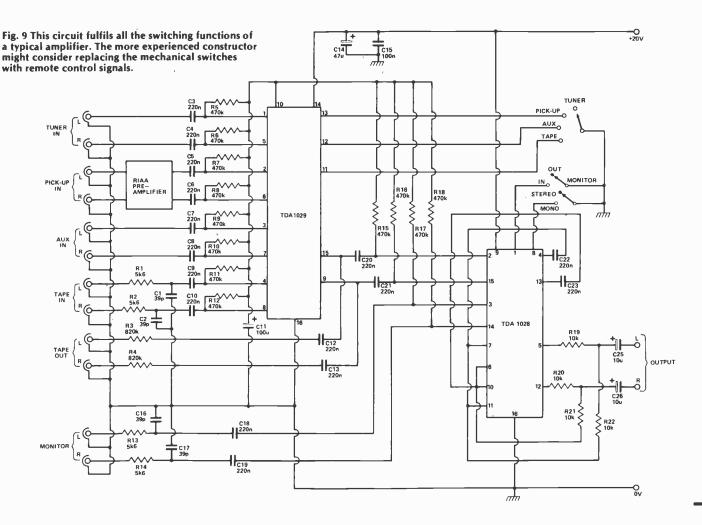
FEATURE: Voltage Controlled Audio

Fig. 8 Stereo signal-source switch for a hi-fi system. The control inputs could be interfaced with a remote control system.





Our prototype board for the TDA 1029. Once again, the optional low-pass capacitors have been omitted.



FEATURE : Voltage Controlled Audio

Similarly, the TDA 1029 can be used in filter modes, taking control of four separate filters; linear, subsonic, scratch and mute in the circuit of Fig. 11. Table 2 shows that the control pins have been chosen so that the mute signal has an overriding effect on all the other control signals (as it should), and the rumble filter signal overrides that of the subsonic filter.

In conclusion, the previous applications show the TDA 1028 and TDA 1029 to be very versatile devices. The simplicity and ease with which they can be used means that they will be popular. Both chips are available from Ambit International.

Connected nine	Control voltages			
Connected pins	Pin 11	Pin 12	Pin 13	
1 — 15, 5 — 9	Н	Н	Н	
2 — 15, 6 — 9	Н	Н	l L	
3 — 15, 7 — 9	Н	L	н	
3 — 15, 7 — 9	Н	L	L	
4 — 15, 8 — 9	L	Н	Н	
4 — 15, 8 — 9	L	Н	L	
4 — 15, 8 — 9	L	L	Н	
4 — 15, 8 — 9	L	L	L	

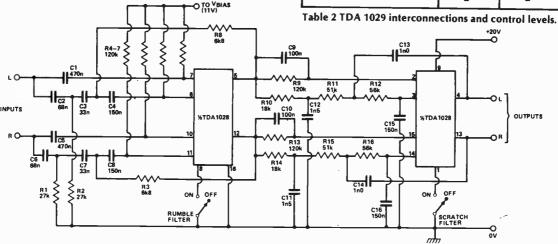
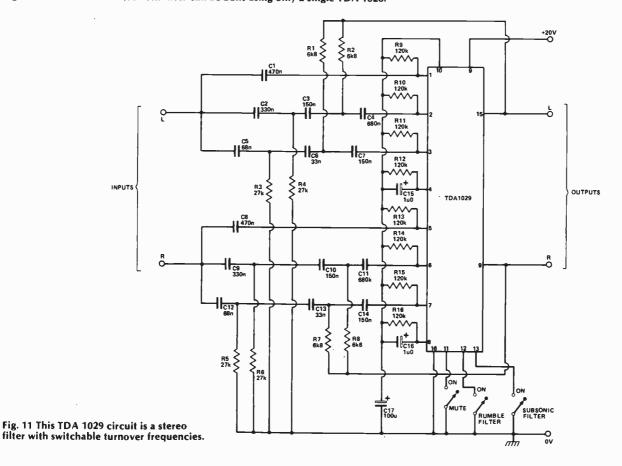


Fig. 10 This stereo rumble/scratch filter can be built using only a single TDA 1028.



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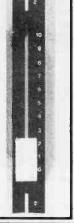
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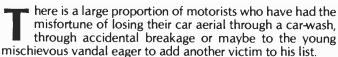
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ANTENNA EXTENDER

If you intend to buy (or already own) a motorised antenna then you shouldn't be without this intelligent gadget which automatically controls your aerial's ups and downs. Design by Ray Marston. Development by Steve Ramsahadeo.



Judging by the number of coathangers that have found their way out of the family wardrobe and into the orifice where the chrome rod once stood, one only has to wait anxiously for the coathanger industry to replace the car aerial. It is not surprising that with these catastrophic events prevailing, drivers are reluctant to replace their aerial. However, if you install a motorised antenna incorporating the ETI antenna controller, it will reduce the risk of losing your aerial.

The unit is designed to replace the manual operation of 'holding' the antenna switch in the 'on' position to activate the aerial. The ETI controller overcomes this hindrance by sensing whether the radio is 'on' or 'off' state and automatically extending or retracting the aerial.

Better Safe Than Sorry

There are also certain fail-safe features which are incorporated to comply with the manufacturer's instructions. These are:

- (i) when the antenna has extended it should not be switched from up to down or vice versa without waiting for at least 3 s before the next operation.
- (ii) switching the radio on and off repeatedly will have no effect while the aerial is operating.

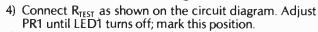
With these features our project supersedes most commercial units already available.

Construction And Setting Up

Construction is straightforward. All components, including the relays, are mounted on a single PCB. Begin construction by inserting all low profile components, ie wire links, Veropins and sockets, followed by resistors, diodes, capacitors and transistors, observing the orientation of all polarised components. R15 is soldered underneath the PCB between the junction of PR2/PR3, and the positive end of C5.

Before you fit the PCB in its box the following setting-up procedure should be carried out:

- 1) Fit IC1 and link points B and C. (This link is used for setting up only.)
- Connect a 12 V power supply to the PCB supply terminals.
- 3) Adjust PR1 until LED1 just turns on; mark this position.



ANTENNA CONTROLLER

5) Disconnect R_{TEST}. PR1 should now be adjusted to the midway setting of steps 3 and 4.

6) For a final check, R_{TEST} can be reconnected and LED1 will switch on; if all is well the remaining ICs can now be fitted

HOW IT WORKS.

IC1 is configured as a voltage comparator with a fixed reference voltage at pin 3; pin 2 is arranged in the same way except that D1 is included as the sensor. If a load is present (ie the car radio is switched on), the voltage at pin 2 will fall to a value of ($V_{\rm CC}$ -600) mV, this being the forward voltage drop of the diode. This change of voltage is now compared to the reference at pin 3.As the voltage has decreased the output of IC2 will switch to aproximately the supply voltage.

PR1 is incorporated in the circuit to balance the tolerances of R4,5,6 and 7 so that with any extreme changes of voltage or temperature, the comparator will reliably detect a change at pin 2. The output of IC1 is fed to IC2a and b (bilateral switches). These

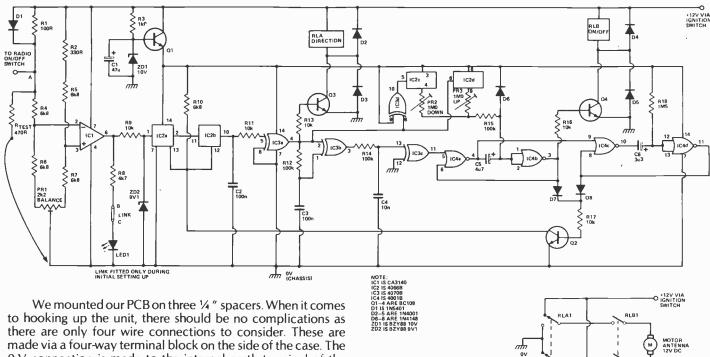
The output of IC1 is fed to IC2a and b (bilateral switches). These switches are normally closed, but with a low signal at their controls (pins 12 and 13) the switches will open, breaking the connection to the rest of the circuit and providing the necessary inhibit facility. ZD1 is added to suppress transients that might cause false triggering

If the car radio is switched on the output of IC2b is high; this voltage is fed to the input of a non-inverting gate (IC3a). The output of this gate determines the relay direction via Q3, as well as providing the input to the edge detector IC3b. The function of this gate is to give a positive-going pulse whenever its input changes state. R14 and C4 are added for protection against spikes that occur during switching. IC3 squares the output of the edge detector, so a reasonably narrow pulse is available to trigger the first monostable (IC4a, IC4b). The output of IC4b energises RLB via Q4 for a period set by PR2 and C5 (this is not more than 5 s). RLB is now supplying power to the motor antenna with a polarity determined by RLA, so while the monostable is turned on the antenna will extend. When this period has ended, pin 4 of IC4a will assume a high state, triggering the second monostable (IC4c, IC4d). The outputs of both monostables are fed to a diode OR gate and inverted by Q2 to open the bilateral switches. This gives a total inhibit time of approximately 6 s, allowing 3 s for the antenna to extend, plus a further 3 s delay before the next operation can take place.

D6 ensures that C5 is fully discharged at the end of the monostable period, to prevent false triggering by residual charge. When the radio is switched off, the output of the comparator

and IC3a will be low. As pin 8 of IC3d is at 0 V, its output will be high which closes IC2c; at the same time IC2d opens and the down sequence is activated. The monostable and RLB follow the same mode of operation as already described.

Q1 and associated components (R3, C1 and ZD1) provide a regulated supply for the CMOS devices.



0 V connection is made to the internal earth terminal of the case. The completed unit can be secured in a suitable place, underneath the bonnet, thus completing the earth return to the battery through the chassis.

Fig. 1 Circuit diagram of the Antenna Extender.

Resistors (all 1/4 W,	5%)	C5	4u7 35 V tantalum
R1	100R	C6	3u3 16 V tantalum
R2	330R		
R3	1k0	Semiconductor	s
R4,5,6,7,10	6k8	IC1	CA3140
R8	4k7	IC2	4066B
R9,11,13,16,17	10k	IC3	4070B
R12,14,15	100k	IC4	4001B
R18	1M5	Q1-4	BC108
		Ď1	1N5401
Potentiometers		D2-5	1N4001
PR1	2k2 miniature horizontal preset	D6-8	1N4148
PR2,3	1M0 miniature horizontal preset	ZD1	BZY88 10V
		ZD2	BZY88 9V1
Capacitors			
C1	47u 16V tantalum	Miscellaneous	
C2	100n polycarbonate	RLA,B	double pole changeover, coil resistance
C3	100n ceramic	ŕ	205R (see Buylines).
C4	10n polycarbonate	Four-way termi	nal block, case ref. BOA 115 (see Buylines).

Fig. 2 Component overlay. Note that R15 is soldered beneath the PCB.

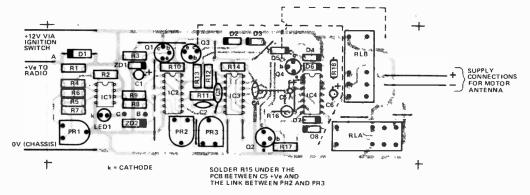
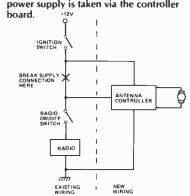
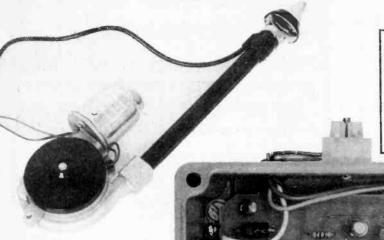


Fig. 3 How to wire up the unit. The radio power supply is taken via the controller





BUYLINES

Grove House Electronics offer a range of motor antennas. The one used in our prototype was a five-section 1 m model NW-510. The relays used are type RL6 from Watford Electronics, and the case is a solid weatherproof, die-cast aluminium housing marketed by West Hyde Developments.

Grove House Electronics, 14 Arcade Chambers, High St., Bognor Regis, Sussex Phone (0243) 861705.

Above: the motor antenna that we used to test our prototype.

Right: the finished PCB mounted in its weatherproof case. The earth terminal for the 0 V connection can be seen in the top left-hand corner. The remaining connections are made via the terminal block on the side of the case.

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1N4001/2 4p	1A/600V* 29p
1N4003 5p 1N4004/5* 6p	2A/50V 30p 2A/100V 32p
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30A/600V	m* 130p
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71 1 1020	345p
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AY-3-8912	965p
5 03 12	748p
AY-5-1224	A
	275p
AY-5-1317	725p
AY-5-4007	
CA3018H	520p
CA3018H	70p 80p
CA3028A	
CA3046N	70p
CA3048	245p
CA3054N	40p
CA3080E	78p
CA3089E	250p
CA3090A0	350p

AY-5-4007D

520p

CA3018H
70p

CA3078
80p

CA3028AH
85p

CA3048
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CA3048
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CA3080E
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CA3130E
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CA3160E
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CA320E

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LM3990N 75p	TDA1024 120p
LM3911 138p	TDA1024 120p
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LM3960 144p	TL071 150p
MC1495E 405p	TL081 45p
MC1495F 80p	TL084 125p
MC3340P 138p	TL170 190p
MK50398 730p	TL170 190p
ML922 485p	VAR2206 340p
ML928 160p	ZN2414 100p
ML928 160p	ZN414 150p
NE531 150p	ZN425E 460p
NE555 26p	ZN1034E 220p
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BC207 13p BC209 13p	BFX87 25p	0110000
BC209 13P	BFX88 25p	2N2369
BC212 10p		2N2484
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BCZ ISL TUP	BFY51 25p	2N2904
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BC261B 23p BC301 32p	BU205 200p	2012052
BC301 32p BC303 32p BC328 17p	BU208 210p	2N3053
BC303 32p		2N3054
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p 74LS210 110p 4098
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78p 4042 85p 4043 99p 4044 30p 4045

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74C98 74C107 74C160 74C161 74C162 74C163 74C192 74C193 74C194 74C195 74C903

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DESIGNER'S NOTEBOOK

CA3140 and LF351 op-amps can be used in some special-purpose single-supply applications that are quite impossible with simple 741-type devices. Ray Marston explains.

ost op-amps can be used with either dual (split) or single-ended supplies, so long as the constraints of their input terminal limits are not exceeded. Specifically, op-amps have a parameter known as 'common-mode input voltage limit' which, in practical terms, defines how close their input terminals can be taken to the positive and negative supply rails without impairing circuit operation. Table 1 illustrates the general input constraints of the 741, CA3140 and LF351 op-amps.

Note here that the input terminals of the 741 op-amp cannot be usefully taken to within closer than a couple of volts of the positive (pin 7) or negative (pin 4) supply rails. Thus the device cannot be used as a true voltage follower in single-supply circuits, for example. The inputs of the CA3140, on the other hand, can swing all the way down to 500 mV below the pin 4 negative voltage, but can only swing within a couple of volts of the positive supply rail. This chip can thus be used as a true voltage follower in single-supply circuits.

Finally, the LF351 inputs can only swing down to within a couple of volts of the negative rail, but can go as high as 100 mV above the positive supply rail. In this way, this chip can be used in some quite unique applications in which input signals are referenced to the positive supply terminals. Let's look at some practical single-supply applications of the CA3140 and the LF351.

OP-AMP TYPE	POSITIVE INPUT LIMIT, REFERENCED TO THE PIN 7 VOLTAGE	NEGATIVE INPUT LIMIT, REFERENCED TO THE PIN 4 VOLTAGE
741	(V+) - 2V	(V-) + 2V
CA3140	(V+) - 2V	(V−) − 500mV
LF351	(V+) + 100mV	(V-) + 2V

Table 1. Common-mode input voltage limits of the 741, CA3140 and LF351 op-amps.

CA3140 Applications

The CA3140 op-amp has PMOS/FET inputs, giving the device a virtually infinite input impedance. The device uses the same pin configuration as the 741 op-amp and can be used with any supply voltage (between pins 4 and 7) from 4 to 36 V. Outstanding characteristics of the device are that its input terminals can swing as low as 500 mV below the pin 4 voltage as already mentioned, and its output can swing to within a couple of millivolts of the pin 4 voltage. A notable defect of the device

is that its output can source far more current than it can sink; when used with a single-ended 5 V supply, it can source 10 mA but sink a mere 1 mA.

Figure 1 shows how the CA3140 can be used as a true voltage follower with a single-ended supply. The input can swing all the way from zero to within 2 V of the positive supply value and the circuit has a virtually infinite input impedance.

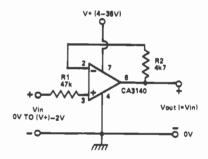


Fig. 1 Zero-referenced DC voltage follower.

Figure 2 shows how the device can be used as a x10 non-inverting DC amplifier that will accept inputs all the way down to 0 V. Again, the circuit has a virtually infinite input impedance.

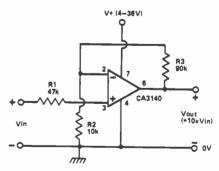


Fig. 2 Non-inverting DC amplifier with a gain of 10.

Figure 3 shows how the above circuit can be modified as a three-range ($100\,\text{mV}-1\,\text{V}-10\,\text{V}$) DC voltmeter or multimeter adaptor with an input impedance of 11 M on all ranges. Offset control PR1 should be trimmed initially to make the meter read correctly at one-tenth of full scale. The two output diodes pro-

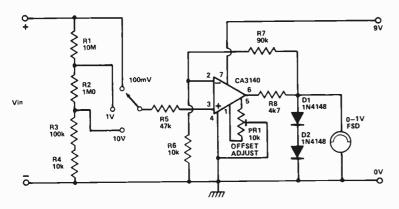


Fig. 3 Three-range DC voltmeter/meter adaptor with 11 M input impedance on all ranges.

tect the meter against damage if excessive input voltages are applied. Note in the Fig. 2 and 3 circuits that the 90k resistor can be made by wiring a 100k resistor in parallel with 1M0.

Figure 4 shows how the basic Fig. 1 circuit can be modified to give a boosted current-driving capacity (up to about 100 mA in this case). Note that the output voltage across emitter resistor R3 is identical to the circuit's input voltage. If Q1 is replaced with a Darlington power transistor, the circuit can easily be made to function as a variable-voltage DC power supply with an output that can swing all the way down to 0 V.

Figure 5 shows how the Fig. 4 circuit can be modified for use as a unity-gain DC level translator that converts a zero-referenced input into an identical positive-referenced output

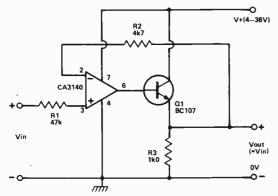


Fig. 4 Zero-referenced voltage follower with boosted output.

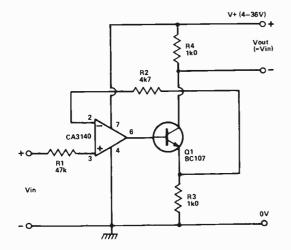


Fig. 5 This unity-gain DC level translator shifts a zero-referenced input to a positive-referenced output.

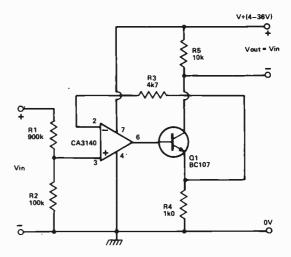


Fig. 6 Unity-gain wide-range DC level translator.

quite independent of supply-line variations. The gain of this circuit is determined by the ratios of R3 and R4, so the circuit can be given a gain of 10(for example) by simply giving R4 a value of 10k. A minor defect of the Fig. 5 circuit is that the output cannot swing below half-supply voltage. Figure 6 shows how the circuit can be modified to enable the output to swing over roughly 85% of the supply voltage range. Here, the input voltage is simply attenuated by 20 dB by R1-R2 and the actual translator is wired in the x10 mode, to give an overall gain of unity.

Figure 7 shows the circuit of a voltage-controlled constant-current generator, with a sensitivity of 10 mA/V. Here, the CA3140 and Q1 are wired as a basic voltage follower, in which the voltage across R2 is identical to the input voltage. Consequently, since the emitter and collector currents are virtually identical, the output current is equal to $V_{\rm IN}/R2$, virtually independent of the value of the output load resistance. The maximum available output current of this circuit is limited by the power-handling capacity of Q1.

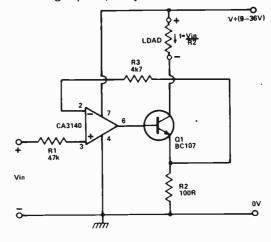


Fig. 7 Voltage-controlled constant-current generator, with a sensitivity of 10 mA/V.

Finally, Figs. 8 and 9 show how the CA3140 can be used as a precision 0-6V8 over- and under-voltage indicator. In both of these circuits the op-amp is used as a simple voltage comparator, with its output feeding to an indicator LED via an 820R limiting resistor; the reference voltage is fed to one input terminal and the sample or test voltage to the other. In the Fig. 8 over-voltage circuit the sample is fed to the non-inverting input terminal, and in the Fig. 9 under-voltage circuit the sample is fed to the inverting terminal.

FEATURE: Designer's Notebook

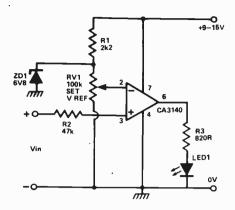


Fig. 8 Precision over-voltage indicator spanning 0 to 6V8.

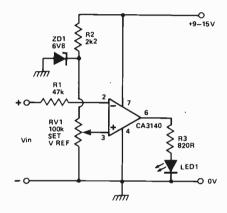


Fig. 9 Precision under-voltage indicator.

LF351 Applications

The LF351 op-amp has JFET inputs, giving the device an input impedance of about a million megohms. The device uses the same pin configurations as the 741 op-amp and can be used with any supply voltage (between pins 4 and 7) from 9 to 36 V. An outstanding feature of the device is that its input terminals can swing as high as 100 mV above the pin 7 positive supply voltage. Defects of the device are that its output can only swing within a volt or two of the positive and negative supply rails and the input can only swing within a couple of volts of the negative rail. A particularly nasty quirk of the IC is that its output inverts if one of its inputs is taken below the negative common-mode limit, or goes high if both inputs are taken below the limit. Nevertheless, the device can be used in some unique single-supply applications in which input signals are referenced to the positive supply terminals.

Figure 10 shows how the LF351 can be used as an overcurrent switch, in which the op-amp output switches high if the current drawn from the supply by an external load exceeds a preset value. Here, a reference voltage of 600 mV is set on the non-inverting terminal of the op-amp by D1, and a currentrelated voltage is applied to the inverting terminal by R_s; the opamp output switches high if the R_s voltage exceeds the 600 mV reference level, so R_s needs a value of 0.6/I, where I is the trip current in amps. Note in the Fig. 10 circuit that the output (into a 10k load) is 1V5 when the switch is off or 1 V below the positive supply when the switch is on. This circuit can be used as a LEDoutput over-current indicator, if required, by replacing R2 with an LED and 820R series resistor.

Figure 11 shows how the LF351 can be used as either a positive-referenced voltage follower or as a voltage-controlled

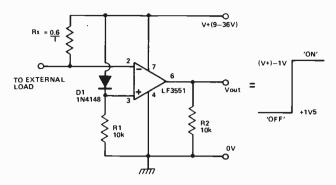


Fig. 10 This over-current switch monitors the current that an external load draws from the positive supply line.

constant-current generator. Here, the op-amp is wired as a standard follower-with-booster-output-stage (using PNP transistor Q1); zener diode ZD1 is used to enable the follower output to swing all the way down to 0 V (referenced to the positive line). The circuit can be used as a constant-current generator by breaking the Q1 collector line as shown, the output current then being equal to $V_{\rm IN}/R2$.

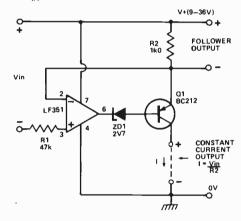


Fig. 11 This circuit can be used either as a positive-referenced voltage follower or as a voltage-controlled constant-current generator.

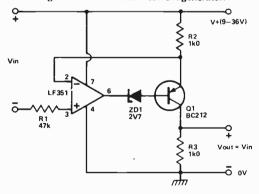


Fig. 12 This unity-gain DC level translator shifts a positive-referenced input to a zero-referenced output.

Finally, Fig. 12 shows how the LF351 can be used as a unity gain level translator that converts a positive-referenced input into a zero-referenced output. The gain of the circuit is determined by the ratios of R2 and R3, so the circuit can be given a gain of 10, for example, by simply giving R3 a value of 10k. Thus if, as an example, the translator is used in the x10 mode and has its input taken from a 1R0 (1 V/A) current-sensing resistor in the positive supply line, the circuit will give a zero-referenced output voltage of 10V/A, thereby simplifying current monitoring problems in power supply circuits.

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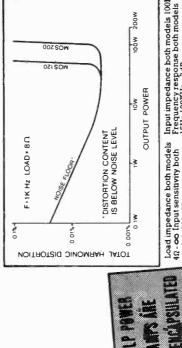
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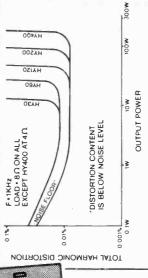
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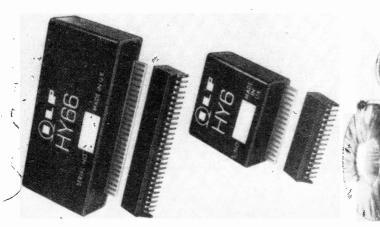


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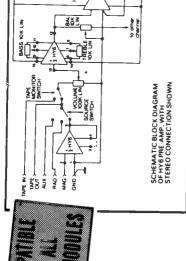
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MINI-DRILL SPEED CONTROLLER

Control the speed of your 12 V minidrill with this inexpensive little unit. No, it's not' one of those oldfashioned PWM controllers so beloved by 'novice' designers, but is a highquality project with built-in feedback speed-regulation and full overload protection. Design by Ray Marston. Project development by

ini-drills are widely used by a whole range of hobby and craft enthusiasts: the electronics hobbyist and engineer find them of particular value in drilling PCBs and deburring holes in panel work, for example. Most mini-drills are designed to run from a 12 V DC supply and draw maximum currents up to about 1 A when heavily loaded or stalled

Steve Ramsahadeo.

The speed of a mini-drill can easily be changed either by varying the DC voltage supply to the drill motor, or by using a simple pulse-width modulation (PWM) technique for feeding pulsed power to the motor. The trouble with both of these simple forms of speed control is that they provide no form of speed regulation.

In practice, the speed of the drill motor falls (sometimes quite dramatically) when the drill is loaded and, simultaneously, the motor input power requirement increases in proportion to the loading level. Since both the conventional 'variable voltage' and the old-fashioned PWM control systems each effectively apply a pre-set voltage level to the motor, they are incapable of meeting the increased power requirements of the loaded motor and the drill speed inevitably falls off as the loading is increased.

ETI's new mini-drill speed controller project overcomes these problems. It uses a 'variable voltage' form of basic speed control but also incorporates a load-sensing feedback network which automatically increases the motor power drive as

loading is increased, to maintain the required drill speed under all loading conditions. The unit is mains powered, has full electronic overload protection and can be used to power any 12 V mini-drill with a current rating below 1.5 A.

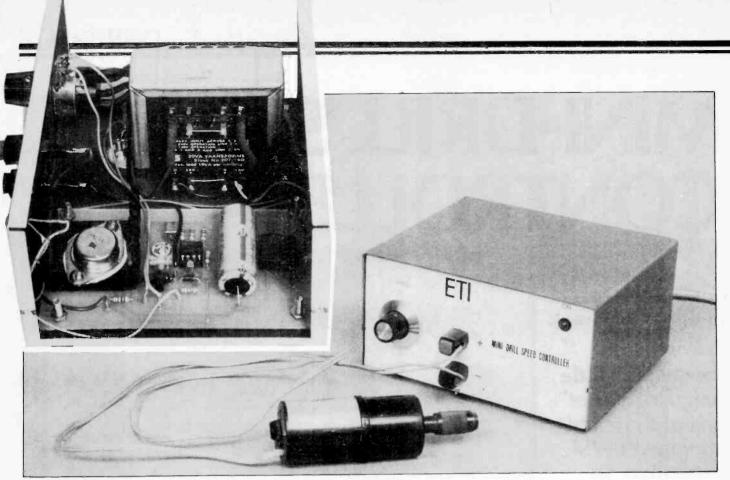
MINI-DRILL SPEED CONTROLLER

Construction

Construction of this project should present very few problems, since all components other than T1, LED1 and RV1 are mounted on a single PCB. Note in the construction that IC2 (a T03-cased regulator chip) must be mounted on a small heatsink and then fixed directly to the PCB.

When the PCB construction is complete, fit the assembly in the specified case, together with transformer T1 and pot RV1, and complete the circuit interwiring. We used a pair of pushbutton loudspeaker terminals for the output — you can use any terminals that suit your drill. The unit can then be given a functional test.

To use the unit, simply connect it to your mini-drill, switch the unit on, and use RV1 to vary the speed of the drill. When using the unit for the first time, set the speed to roughly one-third of maximum, then lightly finger-load the drill chuck and adjust preset PR1 so that the speed remains virtually constant in both the loaded and unloaded states. The unit is then complete and ready for use.



Top: inside the case. Construction is quite straightforward. Above: our prototype unit with a PCB drill connected via the push-button loudspeaker terminals.

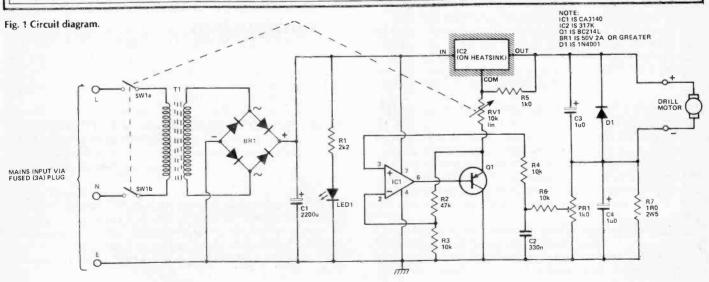
HOW IT WORKS.

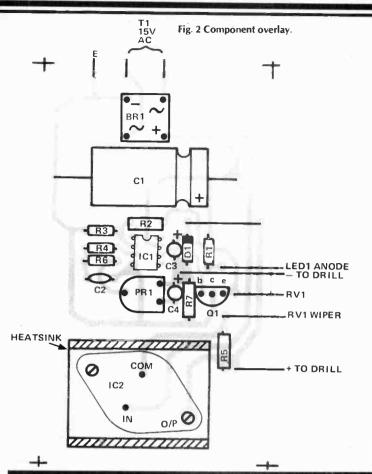
The circuit of the controller contains three basic elements, these being a power supply unit (T1-BR1-C1), a variable-voltage generator (IC2-R5-RV1), and a load-sensing feedback network (R7-PR1-IC1-Q1, etc). The basic (unloaded) speed of the drill motor is determined by the variable-voltage generator circuit; the speed regulation (loaded speed) is controlled by the load-sensing feedback network.

Circuit operation relies on the basic fact that the current consumption of the motor is directly proportional to the motor's loading factor. When the motor is lightly loaded, the current consumption is low; when the loading is high, the current consumption increases. In our circuit, the motor current consumption (and thus the loading factor) is monitored by R7. A fraction of the voltage generated across R7 is tapped off by PR1, filtered by R6-C2, and used to control the variable-voltage generator circuit (IC2) via composite non-inverting amplifier IC1-Q1.

The output of IC2 is determined by the relative ratios of R5 and RV1 and by the voltage on Q1 emitter. When Q1 emitter is at 0 V, the output of IC2 can be varied from 1.2 V to 12 V by RV1. When Q1 emitter is not at 0 V, the Q1 emitter voltage is simply added to the IC2 output voltage. In practice, the Q1 emitter voltage is directly proportional to the load current of the motor. Thus, when the motor loading is increased the resulting increase in motor current causes the Q1 emitter voltage to rise. This increases the output voltage of IC2 to a level sufficient to maintain the motor speed at a constant level. The degree of regulation is controlled by preset PR1.

IC2 is a voltage regulator chip and has built-in short-circuit output protection and thermal overload cut-off. The device thus provides the circuit with a high degree of electronic overload protection and limits short-circuit currents to about 1.5 A, thereby eliminating the need for old-fashioned output protection devices such as fuses.





PARTS LIST.

resistors (arr	4W, 5% except where stated)
R1	2k2
R2	47k
R3,4,6	10k
R5	1k0
.R7	1R0, 2W5
Potentiometer	S
RV1	10k linear with integral mains switch
PR1	1k0 miniature horizontal preset
Capacitors	
C1	2200u 25 V axial electrolytic
C2	330n polycarbonate
C3,4	1u0 35 V tantalum
Semiconductor	rs
IC1	CA3140
IC2	317K
Q1	BC214L
BR1	bridge rectifier; 50 V, 2 A or greater
	1N4001
D1.	0.2" red LED

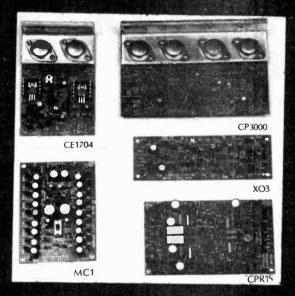
BUYLINES

The 317K voltage regulator can be obtained from Watford Electronics. All the other components are common types and should not present any problems

The case used is available from West Hyde Developments order as Samos 006.

Professional

Crimson modular audio amplifiers feature: Low values of transient and steadystate distortions Envelope distortion (below 500 Hz) less than 0.05% ton-board electronic protection P.C.B. pin and edge connector termination Interpretation of the property components, i.e. P.S.U.'s, beatsinks at a qualitable from Crimson heatsinks etc. available from Crimson.



The Crimson range of amplifier modules are built to very high standards and have earned an enviable reputation in every field to which they have been applied. The boards come ready built and tested (guaranteed for two years) and can be used to advantage where high quality signal amplification is required. The power amplifier modules range from 60WRMS to 310WRMS with up to twice this amount in bridge mode. All feature substantial heatsink brackets which can be botted to any available heatsink or the Crimson purpose designed types. Input sensitivity is set at 775mV and power supply requirements are catered for by one of the three Crimson toroidal power supplies. The Pre-amplifier module (CPR1) sobasically a phono amplifier with sophisticated circuitry incorporating R.I.A.A. equalisation. Also onboard is auxiliary amplification for tape and tuner inputs. A separate module (MC1) is also available and gives the required boost for low output moving coil type cartridges. External components required are potentiometers for volume and balance, switches for signal routing and a regulated ±15V D.C. power source (REG1). Complimenting this range, are the electronic crossover modules XO2/XO3 which, with a special muting board (MU1) can be incorporated in all types of active speaker systems.

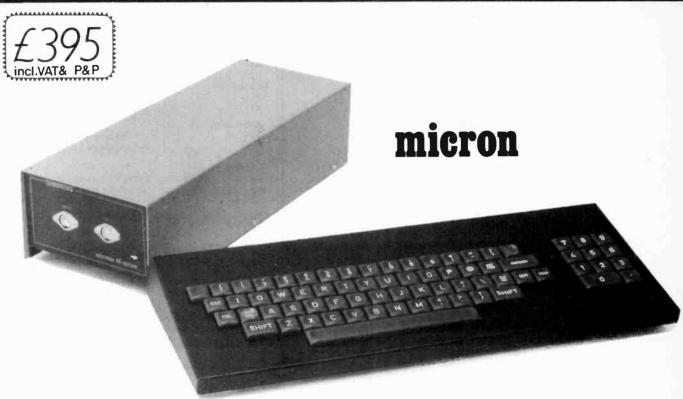
Numerous applications are possible with Ctrimson modules. For example, a complete Hi-Fi Pre & Power

special muting board (MUT) can be incorporated in all types of active speaker systems. Numerous applications are possible with Crimson modules. For example, a complete Hi-Fi Pre & Power amplifier of 40-125WRMS/channel can be built using our Hardware kits (see Hobby Electronics review, August 1980). Alternatively, Mono or Stereo slave amps of up to 500WRMS can be built into proprietory flight cases, while other uses include active loudspeaker systems such as designed by R.I. Harcourt in Wireless World October/November 1980. Further details of how to use the modules are contained in the

SPEC	IFICA	TIONS							T. Carlon	OF STATE
		O/P4ohms*	PSU	H/Sinks	Slew	S/N	Sensitivity	T.H.D. (typ)	F.R	Size
E 608	38		CPS1	50mm	30V/us	110dB	775mV	0 0035%	1 5Hz 50Khz i 3dBi	80 - 120 - 3
E1004	44	70	CPS3	100mm	30 V JuS	110dB	775mV	0 003545	1 5Hz 50Khz i 3dBi	80 - 120 - 1
E1008	65		CPS3	100mm	30V/115	110dB	775mV	0 0035%	1 5Hz 50Khz i 3dbi	BO - 120 -
E1704	85	121	CPS6	150mm/FM1	30V /11S	110dB -	. 775miV	0 0035%	1 5Hz 50Khz (3dB)	BO - 120 -
E1708	125		CPS6	150mm/FM1	30V / S	110dB	. 775mV	0 0035%	1 5Hz 50Khz 3dB	80 - 120 -
P3000		250	CPS6	FM2	JUVAIS	110dB	775mV	0 0035 1	1 5Hz 50K hz i 3d8 i	161 - 102 -
PRIIS	Output	775mV	REGI		3V P : S	70aB	2 BniV RMS	0.008%	20Hz 20Khz	138 - 80
AC1(S)	Output	2mV	REG!			65dB	70uV/150	0.008"	20Hz 20Khz	8U - 120 -
O2 XO3	Output	775 2500 mV	BEG L		9V 115	90dB	775mV	0.01 %	X over points	150 - 50 -
							nodules run using a stab		me power supp .U.	ply. High
		ELD FRO			O AP	RIL '	81 BARC	LAYCARD	Prices include	VAT PE
	mplifier Mod			y Modules		eatsinks				
E 608		21.00 CPS1(8		(119.50	50mm		11 70		anywhere in th	ne UK. CC
E 1004			150VA)		100mm		12 70	'ISA	C . 00	
1008		27 50 CPS612	50VA1	£30 00	150mm		13 50	.5/1	£1.00 extra. (L	,imit E100
E1704		35 00 REG1		19 30	FM1		± 36 00			100
F1708	C:	35 00 TR6		t2 50	FM2	1	142 00		Export Payl	DV ACCES

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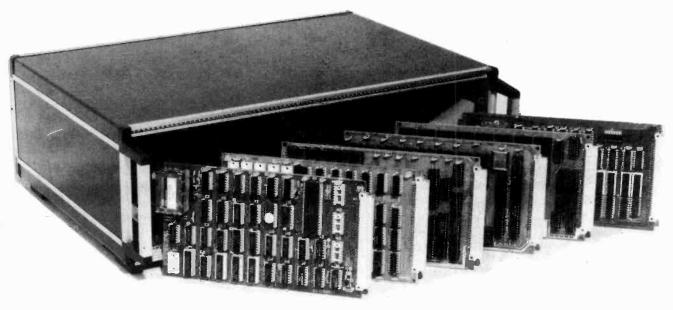
Forehill Works

Forehill Ely Cambs England Tel: (0353)3633

SOFTWARE

- TOK EXTENDED MICROSOFT BASIC.
- ALL THE USUAL BASIC COMMANDS.
- INTEGER AND REAL NUMBERS.
- INTEGER AND REAL ARRAYS.
- INTRINSIC FUNCTIONS: ABS, INT, RND SGN, SIN, SQR, TAB, USR, ATN, COS, EXP, LOG, TAN.
- USER DEFINED FUNCTIONS.
- READ AND DATA STATEMENTS.
- DUMP AND LOAD PROGRAMS.
- PROGRAM EDITING COMMAND.
- STRING FUNCTION FOR TEXT I/O.
- BASIC CAN CALL MACHINE CODE SUB-ROUTINE.
- USER MACHINE CODE INTERRUPT HAND-LER INTERFACES WITH BASIC.
- XBUG.
- DATA CASSETTE FILE HANDLING IN BASIC

microtan 65



The Microtan system is rapidly becoming accepted as the ultimate approach to personal computing. Start with Microtan 65, a 6502 based single board computer, and expand to a powerful system in simple and in-expensive stages. The Microtan system is a concept and not an afterthought, this means expansion is easy and very efficient! Unlike many other systems, you'll find it difficult to outgrow Microtan, and you won't be wasting your money on a product that will only last you a few months! When you are ready to expand, Tanex is waiting. The features offered by Tanex are tremendous, and you can start into them for just £49.45! Cassette interface, 16 I/O lines, two 16 bit counter timers, data bus buffering, memory mapping and a further 1K of RAM are standard. From thereon expansion is simple, just plug in extra integrated circuits to get yourself 8K of RAM, a further 16 I/O lines and two more counter timers a serial I/O line with RS232/20mA loop and full modem control, XBUG - a firmware package containing cassette file handling routines, plus a line-by-line assembler (translator) and dis-assembler, PLUS 10K EXTENDED MICROSOFT BASIC, a suped-up version of the Basic as used by major manufacturers such as Apple, Tandy and Nascom, NO OTHER LOW COST MICROCOMPUTER OFFERS YOU THIS SUPERB PACKAGE. O.K. so you want more memory, try Tanram for size! Upto 40K bytes on one board starting for as little as £50.60. RAM freaks will be pleased to hear that our system mother board offers page memory logic which will support 277K Bytes, satisfied? To house these beautiful modules you can choose between our mini-rack (as used on Micron), which accepts Microtan and Tanex, or our system rack pictured above. The system rack will support 12 modules. What are these extra modules? Well for $\,$ starters there's a couple of I/O $\,$ modules, parallel and serial offering upto 128 I/O lines organised as 16 8 bit ports and 8 serial I/O ports respectively. Shortly we'll be introducing high definition (256x256) colour graphics, A to D and D to A modules, IEEE 488 Bus interface, a PROM programmer, disc controller and TANDOS - a 6502 CPM system. So there's plenty to keep you busy. Send for more details, and find out how you can get started for just £79.35! ALL PRICES QUOTED INCLUDE V.A.T.

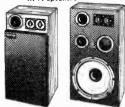
AIM 65, KIM 1, SIM 1 USERS- READ ON!

We have produced a T.V. interface module which simply connects to the expansion socket of your computer and produces a display of 16 rows by 40 characters! Of even more interest will be our Buffer module, which allows you to expand into our system rack, giving you access to the full range of Microtan modules.

Please underline the	information required.
AIM T.V. INTERFACE.	
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ADDRESS:	
PLEASE ENCLOSE 12p ST	TAMP. THANK YOU.
	•

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With controls, attractive woodgrain finish, RMS ratings for 8 ohm.

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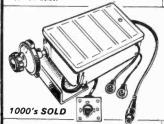
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PROJECT KIT 800104 £26.50

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PROJECT KIT 80502K £25-95 (Musical Box) PROJECT KIT 80543K £3-95 (Stamp Amplifier)

The High COM system of noise reduction, developed by TELEFUNKEN, offers considerable advantages over existing systems. The Compressor/ Expander circuits are designed around a specially designed IC, the U401B.

HIGH COM- NOISE REDUCTION SYSTEM

The unit is designed for use with cassette and reel-to-reel tape systems to give professional quality, low noise recordings.

TECHNICAL SPECIFICATIONS:

Frequency Range 20-18Hz (+0, -3dB) Signal/Noise ratio 80dB (DIN input) 85dB (Socket input)

Distortion Noise reduction less than 0.2% at 100Hz : 15dB

20dB at 3kHz at 15kHz : 25dB

Input Sensitivity DIN 0.6mV/6k ohm Output Sensitivity DIN 130mV/79k ohm

FULL SPECIFICATION AVAILABLE

The project packs include all the electronic components, self adhesive front panel, prebuilt and tested HIGH COM modules and PCBs. The meter circuit uses an LED bar display and is essential for the operation of the unit.

HIGH COM (81117-1)

£49-55

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Case for project

DISCO PROJECTS

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This mixer is easy to assemble, most parts are mounted on the single PCB. It incorporates two stereo disc inputs, two microphone inputs and a stereo tape input. Slider controls for level controls, rotary pots for tone controls. Distortion is approx 0.1%, trequency response 20Hz to 25kHz (3dB) £37-50

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PRÓJECT PACKS include PCB, Components and data, Cases etc



PROJECT PACKS

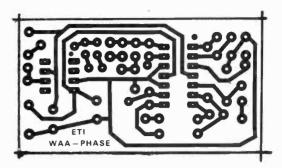
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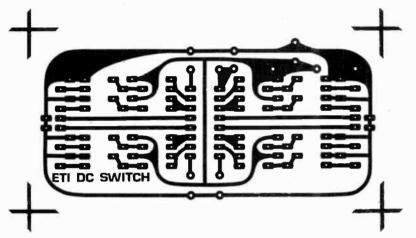
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PCB FOIL PATTERNS

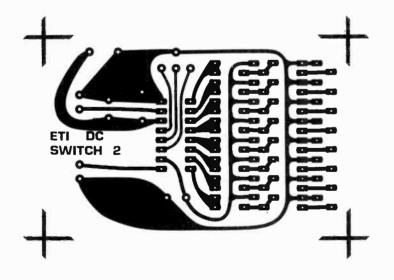


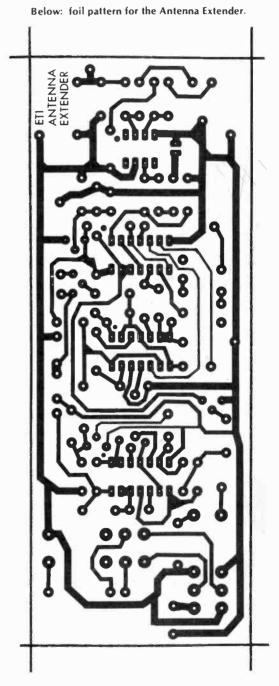
Above: the Waa-Phase PCBs



Above: foil pattern for the TDA 1028 board.

Below: the TDA 1029 printed circuit board.

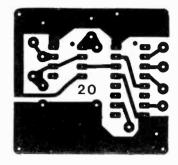






Above: Cross brooch PCB.





Above: the two boards for the Spiral brooch.



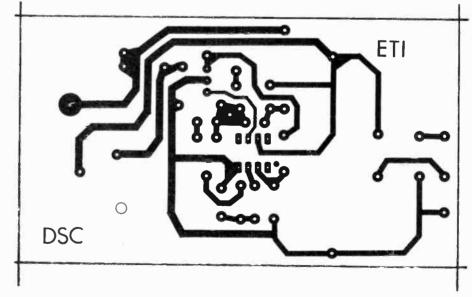


ALIEN ATTACK

Above: the board for Alien Attack.

Below: the Mini-Drill Speed Controller foil pattern.





HIGH POWER MODULE KITS

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4-16 ohms

0.1%

25Hz-20KHz

400mV @ 47K

SPECIFICATIONS

Operating voltage (DC)

Frequency response measured at 100 watts Sensitivity for 100 watts

Typical T.H.O. @ 50 watts 4 ohms load Oimensions

205 x 90 and 190 x 36 mm The P.E. power amp kit is a module for high power applications—disco units, guitar amplifiers, public address systems and even high power domestic systems. The unit is protected against short circuiting of the load and is safe in an open circuit condition. A large safety margin exists by use of generously rated components result, a high powered rugged unit. The PC Board is backprinted, etched and ready to drill for ease of construction, and the

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PRACTICAL ELECTRONICS CAR RADIO KIT (Constructors pack 7)



- * Modern styling design * All new unused components *6 watt output * Ready etched & punched P.C.B.

* Incorporates suppression circuits * Now with tape input socket

All the electronic components to build the radio, you supply only the wire and solder as featured in the Practical Electronics March issue Features Pre-set tuning with five push button options black illuminated tuning scale, with matching rotary control knobs, one, combining on/off volume and tone-control, the other for

manual tuning, each set on wood simulated fascia.

The P.E. Traveller has a 6 watts output, neg ground and incor porates an integrated circuit output stage, a Mullard IF module LP1181 ceramic filter type, pre-aligned and assembled and a Bird pre-aligned push button tuning unit. The radio fits easily in or £10.50 under dashboards

Complete with instructions.

plus £2.00 p&p

CONSTRUCTORS PACK 7A Suitable stainless steel fully retractable locking aerial and speaker (approx 6" x 4") is available as a kit complete f1.95 per pack.



30 + 30 WATT STEREO AMPLIFIER **BUILT AND TESTED**

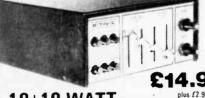
Viscount IV unit in teak simulate cabinet silver finished rotary controls and pushbuttons with matching fascia, red mains indicator and stereo jack socket. Functions switch for mic magnetic and crystal pickups, tape and auxiliary. Rear panel leatures fuse holder. OIN speaker and input. socket 30 + 30 watts RMS 60 + 60 watts peak for use with 4 to 8 ohm speakers Size 14%" x 10" approx

READY TO PLAY £32.90 1380 080

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 Matching ower supply kit with transformer. £3.00 plus £1.96 p8p
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- · Complete with application notes.



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 Multard Stereo Preamplifier module.

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 Converts to a 20 watt Oisco amplifier.

To complete you just supply connecting wire and solder. Features To compiter you just supply connecting while and source. Feature include din input sockets for ceramic cartridge, microphone, tape or tuner. Outputs—tape, speakers and headphones. By the press of a button it transforms into a 20 watt mono disc amplifier with twin deck mixing. The kit incorporates a Mullard LP 1183 pre-amp module, plus power amplifier assembly kit and mains power supply. Also featured 4 slider level controls, rotary bass and trettle controls and 6 push butto switches. Silver finish fascia panel with matching knobs and contrasting ready made black vinyl finish cabinet and ready made metal work. For further information instructions are available price 50p. Free with kit.

SPECIFICATIONS

| Suitable for 4 to 8 ohms speakers | Frequency response | 40Hz - 20KHz | 175mV Aux. 200mV Mic 1 5mV Tone controls | Bass ± 12db @ 60Hz | 176bV = 12db @ 10KHz | 176bV = 176bV

BSR chassis record deck with manual set down and return, complete with stereo ceramic cartridge. £8.50 plus £3.15 p&p when purchased with amplifier Available separately £10.50 plus £3.16 p&p.

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CONVERSION KIT. All components including P.C.B. to convert your ceramic input on the 10+10 amp to magnetic. £2.00 when purchased with kit featured above. £4.00 separately inc. p&p.

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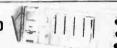
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STEREO DISCO PREAMPLIFIER

matching above modules, suitable output and Mic/Tape input. Ready built, ready to play with circ

diagram and application notes to suit £21.95 plus our power module kits.

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50 WATT MONO DISCO AMPLIFIER

Size appox 13%" x 5%" x 6%". 50 watts rms. 100 watts peak output. Big features include two disc inputs, both for ceramic cartridges, tape input and microphone input. Level mixing controls fitted with integral

push-pull switches Independent bass and master volume



BSR Manual single play record deck with auto return and cueing lever. Fitted with stereo ceramic cartridge 2 speeds with 45

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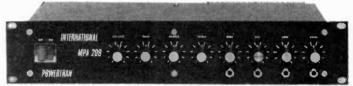
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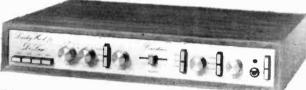
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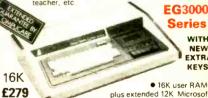
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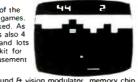
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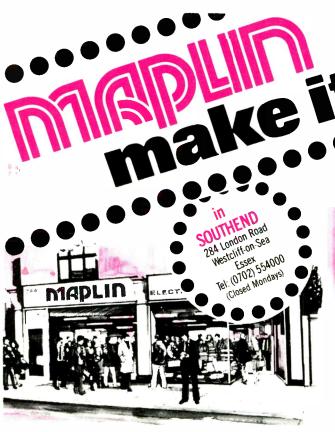
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