CICITONICS
APRIL 1977
TOday international 35p



HIGH YOUR OWN MICHAELMPUTER WITH System THE MANY

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ALSO INSIDE 5 YEAR INDEX ALARM CONTROL FUZZ BOX BENCH POWER SUPPLY

NEWS. . . CONSTRUCTION . . . DEVELOPMENTS . . . AUDIO

	<u>nn</u>	n E	FOTDOL	HAC	TTL	74			_			_
■TALE	UK		LECTRON		7400 7401	14	7475 <b>36</b> 7476 <b>36</b>	74157 74159		9AE 43 LINE		*C1304P <b>360</b>
			ORD, HERTS., E		7402 7403	16 16	7480 <b>50</b> 7481 <b>114</b>	74160 74161	116 402 116 402	1AE 105 702 2AE 95 7090	: 14 pin <b>27</b>   8	AC1312PQ 175 AC145BP 77
MAIL ORDER,	CALL	ERS WE	LCOME. Tel. Wat	tford 37774	7404 7405 7406	20 22 44	7482 <b>82</b> 7483 <b>95</b> 7484 <b>95</b>	74162 74163 74164	116 402		8 pin 22 /	AC1496 91 AC1710CG 79 AFC6040 97
			. AND FULLY GUARA RMS OF BUSINESS:		7407 7408	44 22	7485 <b>125</b> 7488 <b>36</b>	74165 74166	120 402 161 402	6AE <b>175</b> 7480 7AE <b>69</b> 753	36 M	MK50253' 550 MM2112N' 450
P.O.s OR BANKERS D INSTITUTIONS OFFICE	RAFT W	VITH ORDE DERS ACCI	R. GOVERNMENT AN EPTED. TRADE AND E	D EDUCATIONAL EXPORT INQUIRY	7409 7410 7411	15 25	7489 <b>390</b> 7490 <b>43</b> 7491 <b>80</b>	74167 74172 74173	590 402	8AE <b>85</b>   8038 9AE <b>109</b>   AY-1- 0AE <b>52</b>   AY-1-	0212 170 1	IE350 160 IE518A 210 IE555 41
WELCOME, P&P ADD POSTAGE AT COST. A	25p° TO	ALL ORDE	RS UNDER £10.00. OV D S.A.E. FOR OUR FR	ERSEAS ORDERS EE LIST.	7412	25 40	7492 <b>53</b> 7493 <b>40</b>	74174 74175	173 403 90 403	3AE 130 AY-5- 4AE 184 AY-1-	1224 349 F 67216 195 F	E556DB 99
Export order	s no VAT.	Applicable to	U.K. Customers only. Unless add 8% to devices mark	s stated otherwise, al	7414 7416 7417	74 35 39	7494 <b>85</b> 7495 <b>73</b> 7496 <b>82</b>	74176	116 404		11 82 1	IE561 395 IE562B 395 IE565A 175
12 ½ %. We stock many more items	s. It pays	to visit us. W	e are situated behind Watf	ford Football Ground.	7420	16 29	7497 <b>262</b> 74100 <b>135</b>	74180 74181 74182	299 404 85 404	4AE 95 CA30 5AE 140 CA30	18 <b>72</b> 1 20 <b>145</b>	IE566' 170 IE567V' 182
Cer Parking space available			gh Street. Open Monday to	Saturday. Ample Free	7422 7423 7425	24 30 30	74104 48 74105 48	7418 <b>4</b> 74185	146 404	6AE 127   CA30 7AE 88   CA30 BAE 58   CA30	28A' 96 F	AM2102-2 <b>390</b> OM2513 <b>750</b> AS560 <b>250</b>
POLYESTER CAPACITORS 400V: 0-001, 0-0015, 0-002 0-047, 0-068, 0-1, 11p; 0-15	2 0.0033	7p; D-0047. (	0-0068, 0-01, 0-015, 0-018	8p; 0-022, 0-033, 9p;	7426 7427	35 36	74107 <b>33</b> 74109 <b>84</b> 74110 <b>54</b>	74188 74190 74191	194 404	9AE <b>52</b> CA30 DAE <b>55</b> CA30	36 147 S 43 165 S	AS570 <b>255</b> G3402 <b>255</b>
150V: 0-039, 0-15, 0-22 11 DUBILIER: 1000V: 0-01, 0-0	p; 0-33, 0-	47 19p; 0-68.	1-0 22p; 1-5 29p; 2-2 32p;		7428 7430 7432	48 18 . 28	74111 <b>76</b> 74116 <b>193</b>	74192 74193	120 405 120 405	4AE 120   CA30 5AE 134   CA30	48 215 8	L414A 220 N72733 125 N76003 240
POLYESTER RADIAL LEAD 0-01, 0-015 5p; 0-022, 0-02	(Values in 7 <b>6p</b> ; 0-03	μf). <b>250V</b> : 33. 0-047, 0-00	58, 0-1 <b>7p;</b> 0-15 <b>10p;</b> 0-22,	FEED THROUGH CAPACITORS	7433 7437	48	74118 <b>115</b> 74120 <b>105</b> 74121 <b>34</b>	74194 74195 74196	95 4060	DAE 112 CA30 PAE 350 CA30	80 <b>80</b> 8	N76013 150 N76023 150
0-33 12p; 0-47 14p; 0-68 1; ELECTROLYTIC CAPACITO	8p; 1-0 21	p; 1-5 <b>24</b> p; 2-	2 28p	1000pF/350V 8p	7438 7440	33 17	74122 <b>50</b> 74123 <b>73</b>	74197 74198	18 406		90AQ 395 S	N76033 235 N76115 215
250V: 100 µF, 40p; 100V: 20 32, 50, 10p; 63, 100, 12p;	0. 6p; 63V 50V: 1-0	6p: 50. 100. 1-	5, 2-2, 2-5, 3-3, 4.7, 6-8 7p; 10p: 220, 10p: 470, 30p; 10	000 43n 2200 48n	7441 7442 7443	74 74 130	74125 <b>89</b> 74126 <b>65</b> 74128 <b>81</b>	CMOS	4070 407 407	DAE 55 CA31 TAE 20 CA31 PAE 20 CA31	30' <b>94</b> T	N76227 175 AA621A 238 AA661A 155
40V: 22, 33, 8p; 100, 10p; 3 80, 100, 160, 8p; 220, 250,	3300, <b>52p</b> 1 <b>3p;</b> 470	; 35V: 33, 7p; 640, 25p; 100	3 30, 470, <b>27p</b> ; 1000, <b>38p</b> ; 00, <b>27p</b> ; 1500, <b>30p</b> ; 2000, <b>3</b>	; 25V: 10, 22, 47, 6p; 4p: 3300, 39p; 4700	7444 7445	122 94.	74132 <b>75</b> 74141 <b>72</b>	4000AE 4001AE	15 4076 15 407	AE 155 LM30 AE 60 LM30	OH 170 T	AA700 <b>253</b> AA960 <b>250</b>
47p; 16V: 10, 40, 47, 68, 6p 540, 10p; 100, 14p. TAG-END TYPE: 70V: 2500	p; 100, 12	5. <b>7p;</b> 470, <b>12</b>	p; 1000, 1500, 18p; 2200, .	25p; 10V: 4 100, 6p;	7446 7447 7448	116 82 80	74142 <b>315</b> 74143 <b>314</b> 74144 <b>314</b>	4002AE 4006AE 4007AE	115 408	IAE 22 LM31	8 <b>155</b> T	AD100 150 AD110 170 BA120S 90
65p; 25V: 4700 48p; 16V: TANTALUM BEAD CAP	4500 38p.	; 40V: 2000+	2000 <b>95</b> p.		7450 7451	17 17	74145 <b>90</b> 74147 <b>275</b>	4009AE 4010AE	55 4510 56 451	AE 45 LM38	1 170 T 9 395 T	BA540Q <b>195</b> BA550Q <b>355</b>
35V: 0.1μF, 0-22, 0-33, 0- 1-0, 2.2μF, 3-3, 4-7, 6-8 25	-47. 0-68. V: 1-5. 10.		METERS (A8 or EGEN) k, 0-25W Log & 0 5W Linear	OPTO ELECTRONICS' LEDs + clip	7453 7454 7460	17 17.	74148 160 74150 128 74151 79	4011AE 4012AE	18 4518	3AE 140   LM38 3AE 140   LM39 3AE 130   LM39	00 <b>54</b> T	BA6418 <b>225</b> BA651 <b>160</b> BA800 <b>90</b>
20V: 1-5. 16V: 10 µF, 22, 4-7. 15. 25. 33. 6V: 47 µF	47. 10V:	1KΩ & 2K	Ω (LIN ONLY) Single gang 24p	TIL209 Red 13p TIL211 Grn 28p	7470 7472	32 28	74151 <b>79</b> 74153 <b>82</b> 74154 <b>150</b>	4013AE 4015AE 4016AE	99 4528 55	BAE 140 M252 M253	AA 650 T	BA810S 105 BA820 80
100 uF. 12p each.  MYLAR FILM CAPACITOR	s	5KΩ-2MΩ si 5KΩ-2MΩ si 5KΩ-2MΩ si	ngle gang 24p ngle gang D/P switch 45p ual gang stereo 55p	TIL220 Red 20p .2" Red 16p .2" Amber Green		32 36	74155 <b>76</b> 74158 <b>80</b>	4017AE 4018AE	95	MC72 MC84 MC13	4" 175 T 5P 150 T	BA9200 350 DA2020 300 N414 120
100V: 0-001, 0-002, 0-005, 0-015, 0-02, 0-04, 0-05, 0-0	56 uF 6p	SLIDER PO	TENTIOMETERS	Yellow <b>20p</b> OCP70 <b>40p</b>	TRANS	ISTOR	S			1	031 148 2	T 120
0.1 µF, 0-15, 0-2 7p. 50V: 0.	v	5KΩ-500KΩ	ind linear values 60mm single gang 60p 60mm WS 150 63.50	ORP12 68p 7 Seg Displays	AC117	р 35	BC158		F182' 33	MPSU55 52	TIS91 24	2N3614'169
range 0-5pF to 10,000pF 0-015 uf, 0-022 uf, 0-033 uf	3p . 0-047 μf 4p	PRESET PO	### DOMM WS 150	TIL312 3" C. Anode 155p C. Cath 155p	AC125 AC126 AC127	22 18 18	BC160	32 8	F183" 33 F184" 28 F185" 30	MPSU56 74 MPU131: 39 OC25: 59	ZTX108 12	2N3615'135 2N2663 32 2N3702 11
TETFER Trimmer (Jackson 58p		7	2-2M Mini. Vert. & Horiz. 7p —3-3MΩ Horiz. larger 8p	TIL321 5" C. Anode 165p C. Cth. 165p	AC1281 AC141	18 22	BC168 BC169C	12 E	F194 10 F195 10	OC26' 95 OC28' 58	ZTX300 14 ZTX301 14	2N3703 12 2N3704 11
SILVER MICA (Values in pF	3-3. 4-7.	0 25W 200	Ω-4-7MΩ Vert.    S — Erie make 5% Carbon	DL747 180p Minitrons	AC1411 AC142 AC142K	20	BC171	11 8	F195 14 F197 15 F198 15	0C29 53 0C35 54 0C36 53	ZTX303 22	2N3705 12 2N3706 10 2N3707 11
6-8, 10, 12, 22, 33, 47, 50 82, 85, 100, 120, 150, 220, 330, 360, 390, 600, 820	3, 68, 75, 250, 300. <b>9p</b> each	Miniature Hi	gh Stability, Low noise  GE VAL 1-99 100+	3015 <b>240p</b> 2N5777 <b>54p</b>	AC1761 AC187	18 18	BC177' BC178'	18 B	F200 <b>37</b> F224A <b>15</b>	0C41' 13 0C42' 13	ZTX311 15 ZTX314 24	2N3708 9 2N3709 9
1000, 1800, 2000, 2200 CERAMIC TRIMMER CAPA	12p each	0.25W 2 2Ω-4.1	7M E24 1.5p 1p	SWITCHES' TOGGLE 2A, 250V	AC188* ACY17 ACY18	19 35 28	BC182	10 B	F244 <b>34</b> F256' <b>34</b> F257' <b>34</b>	0C44° 25 0C45′ 13 0C46′ 35	ZTX500 17	2N2710 12 2N3711 12 2N3771'164
2-7pF 4-15pF 6-25pF B-30p	pF 20p	0.5W 2.2Ω-4 1W 2.2Ω-10	7M E12 2p 1.5p	SPST 23p DPDT 29p 4 pole on/off 35p	ACY19 ACY20	25 24	BC183L	11 B	F258' 38 F259' 48	0C70' 35 0C71' 23	ZTX502 19 ZTX503 19	2N3772*175 2N3773*275
MINIATURE TYPE TRIMMI 2 5-6pF; 3-10pF; 10-40pF 5-25pF, 60pF; 88pF, 190F	ERS 22p 30p	2% 0 5W M	etal Film E12 8p	SUB-MIN TOGGLE SP changeover 48n	ACY21 ACY22 ACY28	29 16 21	BC+84L	12   B	F594 30 F595 28 FR39 25	OC72' 27 OC77' 56 OC79' 53	ZTX504 47 ZTX531 28 ZTX550 18	2N3819 22 2N3820 42 2N3823 49
COMPRESSION TRIMMER	S	(Practical Will Complete Kit	GENERATOR reless May-June 1976) (Inc VAT) £42,95	SPST on/off 77 44p DPDT 6 t 69p DPDT Centre off 88p	ACY39 ACY40	78 24	BC187 BC212	23 B	R40 <b>25</b>	OC810 24	2N526 40 2N696 15	2N3824: 39 2N3866: 90
3-40pF, 8-80pF, 20-140pF 50-200pF, 100-500pF, 1250	25p pF 33p	Ready built (	incl VAT) £52.95 r complete list	SLIDE 250V:	ACY44 AD149 AD161	39 46 38	BC213	11 B	R80 25 X18' 54 X29' 28	OC827 25 OC82D 35 OC83 28	2N697 21 2N698 30 2N699 45	2N3903 15 2N3904 15 2N3905 17
JACK PLUGS	Plastic		SOCKETS	1A DP C/O 12p	AD162' AF114'	36 20	BC214 BC214L	14 B	X38' <b>38</b>	OC84" 44 OC84N 44	2N706' 15 2N706A' 18	2N3906 17 2N4037 39
chrome 10p	body <b>8p</b>	open metal 8p	moulded in line couplers break 11p	4 pole / 2-way 15p PUSH BUTTON Spring loaded	AF115' AF117'	20 20 23	BC328	18 B	X55 40 X64 38 X81 141	0C123° 115 0C139° 140 0C140° 125	2N707' 50 2N708' 15 2N914' 18	2N4058 15 2N4061 13 2N4289 24
3.5mm 14p MONO 19p STEREO 28p	10p 15p	8p 13p	contacts 12p 20p 18p	SPST on / off 55p SPOT C / over 65p	AF118' AF121'	47 33	BC462	30 B	X84 24 X85 28	OC141' 157 OC170' 30	2N916 27 2N918 34	2N4859 <b>35</b> 2N4871 <b>34</b>
STEREO 28p	18p PLUGS	\$OCKETS	22p 22p SWITCHES * PUSH BUT	OPDT 6 Tag 85p	AF124 AF125 AF126	30 30 30	BC548	15 B	X86 28 X87 23 X88 26	0C171' 32 0C201' 125 0C202' 135		2N5135 12 2N5136 12 2N5138 12
2 PIN Loudspeaker 3, 4, 5 (180 & 240 )	12p	8p	Miniature Non Locking Push to Make 15p	Push to Break 25p	AF127	30 33	BC557 BCY30	28 B 55 B	Y50 17 Y51 17	OC203 150 OC204 150	2N1131' 19 2N1132' 23	2N5179 60 2N5180 60
CO-AXIAL (TV)	14p	10p	ROCKER (white) 10A 250 SP changeover centre off ROCKER: (black) on/off 1	25p	AF178' AF179' AF180'	70 70 70	BCY34	80 B	Y52' <b>17</b> Y65' <b>25</b> Y26' <b>40</b>	TIP29 43 TIP29A 45 TIP29C 68	2N1302° 25 2N1303° 27 2N1304° 45	2N5191 65 2N5305 40 2N5457 35
PHONO assorted colours	9p	5p (Single, 7p (Double)	ROCKER: Illuminated (whi	(te)	AF181' AF185'	48 60	BCY40" BCY55" 2	90 B: 04 B	Y29' <b>85</b> Y78 <b>25</b>	TIP30 52 TIP30A 56	2N1305° 32 2N1306° 37	2N5458 36 2N5459 36
Metal screened  BANANA 4mm	12p	10p (Triple)	ROTARY: (ADJUSTABLE : way. 2p/2-6W, 3p/2-4W, ROTARY: Mains 250V AC	STOP) 1 pole / 2-12 4p/2-3W <b>30p</b>	AF186 AF239 ASY26	48 39 38	BCY70	17 B	SX20" <b>20</b> SY95A <b>18</b> J105' <b>195</b>	TIP30B 64 TIP30C 74 TIP31 52	2N1307' 37 2N1308' 48 2N1671'150	2N5485 42 2N6027 35 40311 38
2mm 1mm	13p 11p	13p 11p	DIL SOCKETS* (Low Prof	ile — Texas)	ASY29' BC107	40	BCY72' BO115'	15 E:	567 <b>48</b> D8001	TIP31A 54 TIP31B 58	2N1671B 190 2N1893 27	40313 93 40316 60
TRANSFORMERS* (Mains	Prim 220-	_	8 pin <b>10p</b> ; 14 pin <b>12p</b> ; 1 <b>30p</b> ; 28 pin <b>45p</b> ; 40 pin	7р.	BC107B BC108 BC108B	9	B0123		158 J400' 90 J491' 160	TIP31C' 63 TIP32' 60 TIP32A' 63	2N2160° 30 2N2217° 40 2N2218A°25	40317' <b>38</b> 40326' <b>40</b> 40327 <b>48</b>
9-0-9 v 75mA 95p	15-0-15V 8-0-18V	A 275p+	PANEL METERS * Full so 59 × 46 × 35mm req 11/2 0-50 µA 0-100 mA	hole	BC108C	12	BD131'	36 M	J2 <b>955'125</b> JE340' <b>45</b>	TIP32B' 80 TIP32C' 83	2N2219A*24 2N2220A*26	40347 <b>65</b> 40348 <b>73</b>
12-0-12V 100mA 98p 15-0-15V 100mA 185p	30-0-30V 6-0-6V 1-	1A <b>295p+</b> 5A 7 345p+	0-100 µA 0-500 mA 0-500 µA 0-1 Amp	50 uA-0-50 uA 100 uA-0-100 uA 500 uA-0-500 uA	BC109B BC109C BC113		BD135	45 M	JE370' <b>68</b> JE371' <b>80</b> JE520' <b>65</b>	TIP33" 95 TIP33B" 112 TIP33C" 120	2N2221A*21 2N2222* 21 2N2303*250	40360° 43 40361° 43 40362° 43
0-6 0-6V 280m A <b>150p</b> 0-12 0-12V 0-5A <b>240p</b>	9-0-9V 2A 30-25-20 2A	270p+ -0-20-25-30 497p+	0-1mA 0-50V 0C 0-5mA 0-300V 0C	3.25 each EDGEWISE	BC114 BC115	16 18	BD137 BD138	42 M 47 M	JE521 <b>74</b> JE2955 <b>98</b>	TIP34' 110	2N2369' 14 2N2483' 30	40411 220 40412 42
15-0-15V 0-5A <b>220p+</b> 20-0-20V 6VA <b>220p+</b>	LT44 Min 20K. Sec.	Driver Prim. 1K 52p	0-10mA S 0-50mA VU''' Price £3.55 each	89×32×70mm 0-1mA_0-500µA Price <b>£4.95</b>	BC116 BC117 BC118	19 20 15	80140	57 M	JE3055' <b>82</b> PF102 <b>36</b> PF103 <b>36</b>	TIP348 140 TIP35 219 TIP35A 225	2N2484° 30 2N2614 54 2N2646° 41	40476 <b>170</b> 40494 <b>89</b>
9-0-9V 1A 245p+	1 2K, Sec MOT Min.	n. O/P Pri. : 3-2Ω 54p : o/P Pri	108 × 82 5 × 38mm reg. 0-50 µA, 0-100 µA, 0-500 µ	60mm panel hole A £5.30	BC119 BC135	25 13	BO145' 19	55 M	PF104 <b>36</b> PF105 <b>36</b>	TIP35C" 270 TIP36A" 346	2N2904A*24 2N2906* 18	40495 <b>93</b> 40594 <b>98</b> 40603 <b>53</b>
30-24-2015-12-0 1A Multi tappings 360p+	1 2K, Sec		COPPER CLAD BOARDS	A approx.) £5.25	BC136 BC137 BC140	15 15 34	BOY61: 1:	15 M	PF106 50 PF107 50 PSA05 24	TIP41A' 66 TIP41B' 73 TIP42A' 78	2N2907' 20 2N2907A' 22 2N2926G 10	40636 110 40673 53
30-24-20-15-12-0 2A multi tap 445p+			SRBP 7 75" × 8 25" Fibre Glass 6 × 6" 6 × 12"	48p 64p	BC1421 BC1431	34 24 24	BF154' 2	22 M 29 M	PSA06 <b>24</b> PSA55 <b>24</b>	TIP428' 82 TIP2955' 85	2N29260 8 2N2926R 3	Matched Pair
(Please add 48p p&p charge t marked + above our normal p			FERRIC CHLORIDE* 11b t	115p	BC147 BC148	7	BF173' BF177'	25 M 26 M	PSA56 <b>24</b> PSA70 <b>26</b>	TIP3055' 50 TIS43 27	2N2926Y 3 2N2011: 30	10p extra
KNOBS * fit 1/4" shaft with gri except K2 (push fit) & KB (f	ub screws		Anhydrous 65p + 30p p  DALO ETCH RESIST	REGULATORS*	BC149 BC153 BC154	9 99 20	BF179'	33 M	PSU02 60 PSU05 46 PSU06 52	TIS46 40 TIS50 47	2N3053' 16 2N3054' 49 2N3055' 49	
K1 Black or White pointer type K2 Slim silvered aluminium	•	9p 10p	PEN★ + spare tip 7	75p TBA625B 95p TO3 Can	BC157	10			SU52 60	TIS74 47	2N3525 76	
K4 Black serrated Metal top w line indicator 33m diam. K4A As above but 25mm diam		22p 20p	VEROBOARD ★ Pitch 01 0.15 0	1A 5V 170p 1A 12V 180p 1A 15V 190p	DIODES					ZENERS Rng 3 3V-33V		
K5 Black fluted metal top and calibrated 0-10 37mm diam	skirt n	26p	(copper clad) (pl 2½ × 3¼" 36p 29p	lain) 1A 18V 210p 19p LM309K 150p	AAZ15 AEY11	15 60	OA70 OA79	8 INS	16 5 001/2 5	400MW 9p 1 3W 17p	RECTIFIERS	2A100V 44 2A200V 46
K6 PK2 as K5, pointer on skir K7 Black, knurled, tapered Me top & skirt Calib. 0-10, 30n	tal	26p 25p	3¾ × 3¾" 43p . 40p	24p LM323K 625p — 1A-5V 220p 34p 1A-12V 220p	BA 100 BA 102	10 10	OA81 OA85	10 IN4	003° 6	VARICAPS MVAM2 135p	(plastic case) p 1A50V 21 1A100V 24	2A400V 56 4A100V 72 4A400V 79
K8 Black or silvered for slider K9 Solid alum. Amplifier ki	pot nob. Profe	10p ssional type.	4½ × 17"134p 107p 7 3¼ × 17"173p 143p 9	73p Plastic	BY100 BY126 BY127	24 12 12	OA90 OA91 OA95	6 IN4	006/7' <b>7</b> 148 <b>4</b> 100V' <b>18</b>	88104 <b>40p</b> 88105B <b>30p</b>	1A200V 26 1A400V 31	4A800V 120 6A100V 73
with etch line indicator 16 5 K10 As above tapered 18½ x	5 × 12mm 17mm	diam. 26p 28p	4% × 17" 222p 14  Pkt of 36 pins 2	45p 1A 12V 130p 28p 1A 15V 135p	0A9 0A10	40	OA200 OA202	9 3A 8 3A	400V 20 600V 27	BB106 45p		6A3400V 92 BY164 56
K11 Alum (top hat) Knurled 11	omm skirt,	noticined 30p		1A 24V 140p	OA47	8	IN914		1000V 30	Noise Diode Z5J 105p		

# ACCOUNTED TO A SINTERNATIONAL INTERNATIONAL

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# PAKS - PARTS - AUDIO MODULES

### **PANEL METERS**

4" RAN	GE		
Size 41/4" >	31/4" x	1 3/4"	
Value	No.		Price
0-50UA	1302		£4.50
0-100UA	1303		£4.50
0-500UA	1304		£4.50
0-1 MA	1305		£6.00
0-50V	1306		£6.00

2" RAN	GE		
Size 2%" x	13/4" x	11/2"	
Value	No.		Price
0-50UA	1307		£3.50
0-100UA	1308		£3.50
0-500UA	1309		£3.50
0-1MA	1310		£3.50
0.50V	1311		£3 50

MR2P	TYPE	
Size 42x4	2x30mm	
Value	No.	Price
0-50UA	1313	£4.80
0-1 MA	1315	£3.20

EUGE	VISE	
Size 31/2"	x 136" x 214"	
Cut out 2	¾′′ x 1¼′′	
Value	No.	Price
0-1MA	1316	£4.05
0-500UA	1317	£4.05

EDGEW/ICE

#### MINIATURE BALANCE/TUNING

IVIETER	
Size 23x22x26mm	
Sensitivity	
100/0/100MA	
No.	Pric
1318	£1.9

#### BALANCE/TUNING

Sensitivity	
100/0/100UA	
No.	
1319	

### MIN. LEVEL METER

Price £2.00

Sensitivity 200UA	
No.	Price
1320	£1.95

Vu METER
Size 40x40x29mm
Sensitivity 130UA

No. 1321	,	Price £2.00
1321		£2.00

#### MINI-**MULTI-METER**

Size	60x24x	90mm			
	sitivity 10				
AC	VOLTS	0-10,	50,	250.	
100	0				
DC	VOLTS	0-10,	50,	250.	
100	10				
DC CURRENT 0-1-100mA					
D 0 1 FOY - b					

Resistance	0-150K	ohms	
No.			Price
1322			£5.95

#### **HIGH SENSITIVITY** TEST METER

Sensitivity 50,000 Size 61/2" x 41/2"	
AC Volts	0-1.5 to 0/500
DC Volts	in 10 Ranges 0-0.5 to 0/500
DC Current	in 12 Ranges 0-25 ua to 0/10A
	in 10 Ranges
Resistance 0-100	ohms to 0/16 meg in 4 Ranges
Decibels	-20 to 62dB in 10 Ranges

No.	Price
1324	£19.76

	_	P8	P	/	7
Postage otherwit	se s	hown.	Add	extra	nless for

#### **TRANSISTORS**

BI	RAN	D NEV	V — F	ULLY	GUAF	RANTE	ED	
109C	€0.08	Type BC550 BC556	Price '£0.14 '£0.14	BFY52	£0.14	Type TIP2955 TIP3055	Price £0.95	2N37

Į	Type	Price	Type	Price	Type	Price	Туре	Price	Type	Price	Type	Price
ı	AC126	£0.16	BC109C		BC550	'£0.14	BFY52	£0.14	TIP2955	£0.95	2N3708	*£0.07
1	AC127	£0,14	BC147	'£0.09	BC556	'£0.14	BIP19	€0.38	TIP3055	£0.75	2N3708A	'£0.07
1	AC128	£0.12	BC148	°£0.09	BC557	'£0.13	BIP20	£0.38	TIS43		2N3709	'£0.07
1	AC128K	€0.26	BC149	'£0.09	BC558	'£0.12	BIP19/		TIS90		2N3710	'£0.07
1	AC132	£0.15	BC157.	'£0.12	8C559	'£0.14	20 MP	£0.80	UT46		2N3711	'£0.07
1	AC134	€0.15	BC158	'£0.12	BD115	€0.50	BRY39	€0.45	ZTX107	'£0.10	2N3819	£0.20
1	AC137	£0.15	BC159	'£0.12	BD116		BU 105	£1.90	ZTX10B		2N382D	£0.40
1	AC141	£0.18	BC167	'£0.12	BD121	£0.65	BU105/02	£1:95	ZTX109		2N3821	€0.60
1	AC141K	€0.30	BC168	'£0.12	BD123	€0.65	BU204	£1.70	ZTX300		2N3823	€0.40
1	AC142	€0.18	BC169	'£0.12	BD124	€0.70	BU205	£1.70	ZTX500		2N4058	'£0,12
1	AC176	£0.12	BC169C	'£0.12	80131		BU208	€2.40	2N1613		2N4059	'£0.14
i	AC176K	€0.26	BC170	°£0.10	BD132	£0.38	BU208/D2	€2.95	2N1711	€0.20	2N4060	'£0.14
1	AC178	£0.25	BC171	'£0.10	BD131/		E1222	€0.38	2N1889		2N4061	'£0.12
1	AC179	€0.25	BC172	'£0.10	132 MP	£0.80	MJE2955	€0.88	2N1890	£0.45	2N4062	'£0.12
ı	AC180	€0.20	BC173	'£0.12	BD133	€0.60	MJE3055	€0.60	2N1893	€0.30		'£0.18
1	AC180K	€0.30	BC177	€0.16	BD135	€0.36	MJE3440	€0.45	2N2147	€0.75	2N4285	'£0.18
1	AC181	€0.20	BC178	€0.16	8D136	€0.36	MP8113	€0.45	2N2148	£0.70	2N4286	'£0.18
ï	AC181K		BC179	€0.16	BD137	€0.38	MPF102	€0.35	2N2160	£0.80	2N4287	'£0.18
ł	AC187	£0.16	BC18D	€0.25	BD138	€0.45	MPF104	€0.39	2N2192	€0.38	2N4288	'£0.18
1	AC187K		BC181	'£0.25	BD139	€0.54	MPF105	€0.39	2N2193	€0.38	2N4289	'£0.18
ı	AC188	€0.16	BC182L		BD140	€0.60		'£0.20	2N2194	£0.38	2N4290	'£0.18
ı	AC188K		BC183	'£0.10	BD139/			'£0.20	2N2217	£0.22	2N4291	'£0.18
1	AD140	£0.60	BC183L		140 MP	£1.20		'£0.20	2N2218	£0.22	2N4292	'£0.18
1	AD142	€0.85		'£0.10	8D155	€0.80	MPSA56	'£0.20	2N2218A	€0.20	2N4293	'£0.18
ı	AD143	€0.75	BC184L		BD175	€0.60	OC22	€1.50	2N2219	£0.20	2N4921	'£0.55
1	AD.149	£0.60	BC207	'£0.11	BD176	€0.60	OC23	£1.50	2N2219A	€0.24	2N4923	'£0.65
1	AD161	£0.36	BC208	'£0.11	BD177	£0.68	OC 24	£1.40	2N2904	£0.18	2N5135	'£0.10
1	AD162	£0.36	BC209	'£0.12	BD178	£0.68	OC25	£0.60	2N2904A	£0.21	2N5136	*£0.10
1	AD161/		BC212	'£0.11	BD179	£0.75	OC26	€0.60	2N2905	€0.18	2N5138	'£0.10
1	161 MP	£0.75	BC212L		BD201/		OC28	€0.90	2N2905A	£0.21	2N5194	€0.56
1	AF114	€0.20	BC213	'€0.11	202 MP	£1.70	OC 29	£1.00	2N2906	€0.16	2N5245	'£0.28
ı	AF115	€0.20	BC213L		BD203	€0.80	OC35	£0.90	2N2906A	£0.19	2N5294	£0.34
1	AF116	£0.20	BC214	'£0.12	BD204	€0.80	OC36	£0.90	2N2907	€0.20	2N5296	€0.35
1	AF117	€0.20	BC214L		BD2D3/		OC70	£0.15	2N2907A	£0.22	2N5457	£0.32
1	AF118	€0.40	BC237	'£0.16	204 MP	£1.70	OC71	£0.15	2N2926G	£0.09	2N5458	£0.32
ı	AF124 AF125	€0.30	BC238	'£0.16	BDY20	£0.80	TIC44	'£0.29	2N2926Y	80.03°	2N <b>5</b> 459	£0.38
1		€0.30	BC251	'£0.15	BDX77	€0.90	TIC45	'£0.29	2N29260	.60.08	2N5551	'£0.30
1	AF126 AF127	€0.30	BC251A		BF457	£0.37	TIP29A	€0.44	2N2926R	*£0.08	2N6027	£0.32
1	AF139	£0.32	BC301	£0.30	BF458	€0.37	TIP29B	€0.52	2N2926B	80.03	2N6121	€0.70
1	AF180	€0.58	BC302	€0.28	BF459	£0.38	TIP29C	€0.62	2N3053	£0.16	2N6122	€0.70
1	AF181	£0.58	BC303 BC304	€0.32	BF594	'£0.15	TIP3DA	€0.50	2N3054		40311	£0.36
4	AF186	£0.58	BC304 BC327	£0.38	BF596 BFR39	'£0.17 £0.25	TIP30B	€0.60	2N3055		40313	€0.95
1	AF239	£0.38	BC328	£0.15	BFR40		TIP30C	£0.70	2N3414		40316	€0.58
1	AL102	£0.95	BC328	'£0.15	BFR79	'£0.25	TIP31A	€0.54	2N3415		40317	£0.36
1	AL103	£0.95	BC338	£0.15	BFR80	'£0.28	TIP31B	€0.66	2N 34 16	'£0.29		£0.36
1	AU104	£1.00	BC440	£0.30	BFX29	£0.25	TIP31C	€0.68	2N3417	'£0.29		£0.45
1	AU110	£1.00	BC441	£0.30	BFX30	£0.25	TIP32A	€0.64	2N3614 2N3615		40346	€0.42
1	AU113	£1.00	BC460	£0.38	BFX84	£0.23	TIP328	£0.76	2N3615	£0.90	40347 40348	€0.55
1	BC107A		BC461	£0.38	BFX85	£0.23	TIP32C	£0.80	2N3646		40348	£0.70
J		£0.08	BC47-7	£0.20	BFX86	£0.25	TIP41A	£0.66	2N3702		40360	£0.38
1	BC107C		BC478	€0.19	BFX87	€0.22	TIP41B	€0.70	2N3703	£0.08	40362	£0.38
1	BC108A		BC479	€0.20	BFX88	€0.22	TIP41C	£0.80	2N3704		40406	£0.40
1		€0.08	BC547	'£0.12	BFX90	'£0.55	TIP42A	£0.72	2N3705		40407	£0.28
1	BC108C	80.03	BC548	'£0.12	BFY50		TIP42B	€0.78		10.03	40408	£0.48
1	BC109B	€0.08	BC549	'£0.12	BFY51		TIP42C		2N3707	80.03°	40409	€0.52

#### 74 SERIES TTL ICs

FULL SPECIFICATION GUARANTEED ALL FAMOUS MANUFACTURERS											
Type	Price	Type	Price	Type	Price	Type	Price	Туре	Price	Type	Price
7400	£0.14	7409	£0.15	7441	£0.64	7482	€0.85	7493	€0.40	74122	£0.50
7401	£0.14	7410	£0.14	7442	€0.64	7483	€0.95	7494	€0.88	74123	€0.70
7402	£0.15	7411	€0.23	7445	€0.90	7484	€0.98	7495	£0.75	74141	60.80
7403	€0.15	7412	€0.23	7446	€0.90	7485	€1.20	7496	£0.80	74154	£1.30
/404	€0.15	7413	€0.27	7447	£0.78	7486	€0.30	74100	£1.00	74180	£1.10
7405	€0.15	7414	€0.58	7448	€0.80	7489	€2.90	74110	£0.50	74181	£2.00
7406	€0.30	7416	£0.28	7475	€0.48	7490	£0.42	74118	£0.90	74190	€1.50
7407	£0.30	7417	€0.28	7480	€0.50	7491	€0.75	74119	£1.85	74198	£2.00
7408	€0.15	7440	€0.15	7481	£0.95	7492	£0.45	74121	€0.30	74199	£1.90

# **CMOSICs** Type Price CD4022 £0.95 CD4023 £0.18 CD4024 £0.72 CD4025 £0.18 CD4026 £0.98 CD4027 £0.60

Type Price CD4000 £0.14 CD4001 £0.18 CD4002 £0.18 CD4006 £0.98 CD4007 £0.18 CD4008 £0.95 CD4009 £0.55 CD4010 £0.55 CD4011 £0.18	CD4012 £0.18 C CD4013 £0.50 C CD4015 £0.90 C CD4016 £0.50 C CD4017 £0.90 C CD4017 £1.00 C CD4019 £0.52 C	D4022 £0.95 D4023 £0.18 D4024 £0.72 D4025 £0.18 D4026 £0.98 D4027 £0.60 D4028 £0.85 D4029 £1.15	Type Price CD4031 £2.20 CD4035 £1.05 CD4037 £0.95 CD4040 £0.95 CD4041 £0.82 CD4042 £0.82 CD4044 £0.98 CD4044 £0.98 CD4044 £0.94 CD4045 £1.40	Type CD4046 £1.30 CD4047 £1.10 CD4049 £0.55 CO4050 £0.55 CD4055 £1.20 CD4055 £1.40 CD4059 £0.40 CD4050 £0.40	CD4072 £0.28 CD4081 £0.18 CD4082 £0.28 C04510 £1.30 CD4511 £1.60 CD4516 £1.40 CD4518 £1.08
Type CA3011	Type Price   14304   63.0   14309   14	m Type m C1351P m C1352P m C1352P m C14566 m MC1466L m MC14960 m MC14960 m MC14960 m NE515A m NE540 m NE555 m NE560 m NE561 m NE562B m NE562B m NE5662B m NE5662B m NE5661	'£0.85 UA70	9C 'E0.25 9 'E0.46 'E0.25 00 'E0.40 1C 'E0.30 1 'E0.32 3C E0.50 3 E0.50 1 'E0.20 1 'E0.20 1 'E0.20 1 'E0.20 1 'E0.20 1 'E0.20 1 'E0.20 1 'E0.35	Type Price SN76013N*E1.40 SN76013N*E1.40 SN76110 'E1.50 SN76115 'E1.90 SN76166 'E0.75 SL414A 'E1.75 SL414A 'E1.75 SL414A 'E1.75 TAA550B 'E0.35 TAA6611A 'E1.50 TAA6611A 'E1.50 TAA6611A 'E1.50 TAA6611A 'E1.50 TAA6611A 'E1.50 TAA6611B 'E2.25 TBA641B 'E2.25 TBA641B 'E2.25 TBA641B 'E2.25 TBA641B 'E2.25 TBA8820 'E0.80 TBA8120 'E0.90 TBA82200 'E3.40 TAA270S 'E3.90 'E3.40 'E3.90 'E3.90 'E3.40 'E3.90 'E3.90 'E3.90 'E3.90 'E3.90 'E3.90 'E3.90 'E3.90 'E

#### DIODES

Туре	Price	Type	Price	Type	Price	Type	Туре	Type	Price			
AA129	€0.08	BA173	£0.15	BY127	'£0.16	BYZ13	£0.26	0A85	£0.09	IN34A	£0.07	
AAY30	£0.09	BB104	€0.15	9Y128	£0.16	BYZ16	£0.41	0A90	£0.07	IN914	£0.06	
AAZ13	£0.10	BAX13	€0.07	BY130	'£0.17	BYZ17	£0.36	0A91	€0.07	IN916	€0.06	
AAZ17	£0.10	BAX16	80.03	BY 133	'€0.21	BYZ18	€0.36	0A95	£0.07	IN4148	€0.06	
BA100	£0.10	BY100	€0.16	BY164	£0.51	BYZ19	€0.28	OA182	£0.07	1S <b>4</b> 4	£0.05	
BA102	£0.32	BY107	€0.12	BY176	'£0.75	0A10	€0.35	0A200	\$0.03	15920	£0.06	
BA148	€0.15	BY 105	€0.18	BY 206		0A47	£0.07	0A202	£0.08			
BA154	£0.12	BY114	£0.12	BYZ10	£0.36	0A70	£0.07	SD10	€0.06			
BA155	£0.14	BY 124	'£0.12	BYZ11	£0.31	0A79	£0.07	SD19	£0.06			
RA156	CO 14	9V126	100.15	RY712	ED 31	OAR1	60 O 7	INGA	E0 07			

### SILICON RECTIFIERS

Type 15920 15921 15922 15923 15924 (N4001	Price £0.06 £0.07 £0.08 £0.09 £0.10 £0.05	Type IN4003 IN4004 IN4005 IN4006 IN4007 IS015	Price £0.07 £0.08 £0.09 £0.10 £0.11	Type 1S020 1S021 1S023 1S025 1S027 1S029	Price £0.10 £0.11 £0.13 £0.14 £0.16 £0.20	Type 1S031 1N5400 1N5401 1N5402 1N5404 1N5406	Price £0.25 £0.13 £0.15 £0.16 £0.17 £0.21
IN4001 IN4002	£0.05 £0.06	IS015	£0.09	15029	£0.20	IN5406 IN5407	£0.21 £0.25

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ELECTRONICS
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PRICE£1.80† No. 237 PRINTED CIRCUIT ASSEMBLY PRICE£1.80†

3 amp Volts No 50 THY3A/50 100 THY3A/100 200 THY3A/200 400 THY3A/400 600 THY3A/600 800 THY3A/800

5 Amp Volts No. 50 THY5A/50 100 THY5A/100 200 THY5A/200 400 THY5A/400

5 Amp T Volts No 400 THY5A/400P 600 THY5A/600P 800 THY5A/800P

No. 238 TRANSISTOR POCKET BOOK PRICEE3.90+

No. 225 110 THYRISTOR PROJECTS USING **SCRS & TRIACS** PRICE£2.50†

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> No. 242 ELECTRONICS POCKET BOOK PRICE £3.75†

30 PHOTOELECTRIC CIRCUITS & SYSTEMS PRICE£1.80†

£0.99 £1.22

48 Case

Price £0.54 £0.58 £0.62 £0.77 £0.90 £1.39

#### **THYRISTORS**

Oma TO	18 Case	7 Amp	TO 48 Case
ts No	Price	Volts No.	Price
0 THY600/10	£0.13	50 THY7A/50	£0.48
0 THY600/20	£0.13	100 THY7A/100	€0.51
0 THY600/30	£0.19	200 THY7A/200	€0.57
0 THY600/50	£0.22	400 THY7A/400	€0.62
0 THY600/100	£0.25	600 THY7A/600	€0.78
0 THY600/200	€0.38	800 THY7A/800	£0.92
0 THY600/400	€0.45		
	- 1		
		State of the state	

		10 Amp	•
	1	Volts No	
amp	TO 5 Case	50 THY10A/5	0
olts No.	Price	100 THY10A/1	00
50 THY1A/50	£0.26	200 THY10A/2	00
00 THY1A/200	€0.27	400 THY10A/4	00
00 THY1A/200	€0.28	600 THY10A/6	00
00 THY1A/400	€0.36	800 THY10A/8	00
00 THY1A/600	€0.45		-
800 THY1A/800	€0.58		_

TO 66 Ca

	16 Amp	TO
TO 66 Case Price £0.25 £0.27 £0.33 £0.42 £0.50	Volts No. 50 THY16A 100 THY16A 200 THY16A 400 THY16A 600 THY16A 800 THY16A	1/100 1/200 1/400 1/600
€0.65	30 Amo	70

Volts No 50 THY30A/50

6 8	100 THY30A/100 200 THY30A/200 400 THY30A/400 600 THY30A/600	£1.43 £1.63 £1.79 £3.50
7		
9	No	Price
31	BT101/500R	£0.80
	BT102/500R .	€0.80
	BT106	€1.25
	BT107	£0.93
- 4	BT108	€0.98
1-e	2N3228	£0.70
се	2N3535	£0.70
57	BTX30/50L	£0.33
69	BTX3D/400L	£0.46
B1	C106/4	£0.60

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Please word your orders exactly as printed, not forgetting to include our part number.

#### V.A.T.

Add 121/2% to prices marked Add 8% to others excepting those marked † . These are

P.O. BOX 6, WARE, HERTS

18 BALDOCK ST., WARE, HERTS OPEN 9 to 5.30 Mon/Sat. Tel: 61593

High quality modules for stereo, mono and other audio equipment.



**PUSH-BUTTON** 

Fitted with Phase Lock-loop Decoder

The 450 Tuner provides instant program selection at the touch of a button ensuring accurate tuning of 4 pre-selected stations, any of which may be altered as often as you choose, by simply changing the settings of the pre-set controls.

Used with your existing audio equipment or with the BI-KITS STEREO 30 or the MK60 Kit etc. Alternatively the PS12 can be used if no suitable supply is available, together with the Transformer T538

The S450 is supplied fully built, tested and aligned. The unit is easily installed using the simple instructions supplied

\* FFT Input Stage

- VARI-CAP diode tuning Switched AFC
- Multi turn pre-sets LED Stereo Indicator
- Typical Specification:

Sensitivity 3µ volts Stereo separation 30db Supply required 20-30v at 90 Ma max.

£13.75



magnetic cartridges to be used where facilities exist for the use of ceramic cartridges only It is provided with a standard DIN

input socket for ease of connection. Full instructions supplied.

#### **POSTAGE &** PACKING

Postage & Packing add 25p unless otherwise shown. Add extra for airmail. Min. £1.00

7+7 WATTS R.M.S.



The Stereo 30 comprises a complete pre-amplifier, power amplifiers and power supply. This, with only the addition of a transformer or overwind will produce a high quality audio unit suitable for use with a wide range of inputs i.e. high quality ceramic pick-up, stereo tuner, stereo tape deck etc. Simple to install, capable of producing really first class results, this unit is supplied with full instructions, black front panel knobs, main switch, fuse and fuse holder and universal mounting brackets enabling it to be installed in a record plinth, cabinets of your own construction or the cabinet available. Ideal for the beginner or the advanced constructor who requires Hi-Fi performance with a minimum of installation difficulty (can be installed in 30

TRANSFORMER £2.45 plus 62p p &p TEAK CASE £5.25 plus 62p p & p.

### STEREO PRE-AMPLIFIER

A top quality stereo pre-amplifier

push-button selector switch provides a choice of inputs together with two really effective filters for

high and low frequencies, plus tape

MK. 60 AUDIO KIT: Comprising 2 x AL60's. 1 x SPM80. 1 x BTM80. 1 x PA100. 1 front panel

and knobs. 1 Kit of parts to include

and knobs. I kit of parts to include on/off switch, neon indicator, stereo headphone sockets plus instruction booklet. **COMPLETE PRICE £29.55** plus 85p postage.

Comprising: Teak veneered cabinet

parts include aluminium chassis heatsink and front panel

TEAK 60 AUDIO KIT:

size 1634"x111/2"x334"

bracket plus back panel

and appropriate sockets etc. KIT PRICE \*\*\*

plus 85p

postage

tone control unit. The six



equency Response + 1dB 20Hz 20KHz. Sensitivity of inputs Tape Input 100mV into 100K ohms

Radio Tuner 100mV into 100K ohms Magnetic P.U. 3mV into 50K ohms

P.U. Input equalises to R1AA curve with 1dB from 20Hz to 20KHz Supply — 20-35V at 20mA.

Dimensions 89mm v

AUDIO AMPLIFIER MODULES

The AL20 and AL30 units are similar in their appearance and in their general specification. How ever, careful selection of the plastic power devices has resulted in range of output powers from 5 to 10 watts R.M.S.

The versatility of their design makes them ideal for use in record players, tape recorders, stereo amplifiers and cassette and cartridge tape players in the home

■ Harmonic Distortion Po=3 watts f=1KHz 02.5 %

 Load Impedance 8-16ohm • Size: 75mm x 63mm x 25mm Frequency response ±3dB Po=2 watts 50Hz-25Hz

Sensitivity for Rated O/P — Vs=25v. RL=8ohm f=1KHz 75mV.RMS

AL20 5w R.M.S. £2.95

other

SPECIFICATION:

AL30 10w R.M.S. £3.25



25 Watts (RMS)

\* Max Heat Sink temp 90C. \* Frequency response 20Hz to 100KHz \* Distortion better than 0.1 at 1KHz \* Frequency Response 20Hz-20KHz (-3dB). Bass and Treble range 20Hz to 100KHz \* Distortion better than 0.1 at 1KHz \* frequency nespons 7 (-3dB). Bass and Treble range (-3dB). Bass and Treble range 12dB. Input Impedence 1 meg ohm. Design Improvements \* Load — 3,4,8, or 16 ohms \* Imput Sensitivity 300mV. Supply requirements 24V.5mA. Size 152mm. 13mm.

Especially designed to a strict specification. Only the finest components have been used and the latest solid-state circuitry incorporated in this powerful little amplifier, which should satisfy the most critical A.F.

PA12, SA450 etc

Modules. Features include on/off volume.

Balance, Bass and Trebie controls. Complete

Power supply for AL20/30,

#### Stabilised Power Supply Type SPM80

SPM80 is especially designed to power 2 of the AL60 Amplifiers. up to 15 watts (R.M.S.) per channel simultaneously. With the addition of the Mains Transformer BMT80, the unit will provide outputs of up to 1.5A at 35V. Size: 63mm. 105mm. 30mm Incorporating short circuit protection.

Transformer BMT80 £2.60 + 62p postage

**OUR PRICE** Input voltage 15-20v A.C. Output voltage 22-30v D.C. Output current 800 mA Max. Size 60mm x 43mm x 26mm. £1\_30 Transformer T538 £2.30



P.O. BOX 6. WARE, HERTS.

**NEW PA12 Stereo** Pre-Amplifier completely redesigned for use with AL 20/30 Amplifier

18 BALDOCK ST., WARE, HERTS SHOP OPEN 9 to 5.30 Mon./Sat. Tel. 61593 AT

-nerve digest

#### **DISPLAY OF FLEXIBILITY**

O.K. so what is it? Briefly, it's a 4 function calculator with 8 digit green display. Additionally, it has a 4 position programme switch. Set it to 'Time' and touch the 'Time' button and it displays day of month, hours and minutes. At the same time it indicates day of week and seconds by flashing periods. (If the machine is operating from the adaptor it will continue to display time until you touch 'Time' again.) Move the slide to 'ST' (for stopwatch) and touch the 'Time' button and it becomes a 10th of a second stopwatch. Touch the 'LAP' button and it holds the display whilst continuing to count out of display - touch the 'LAP' button a second time and it picks up elapsed time in display.



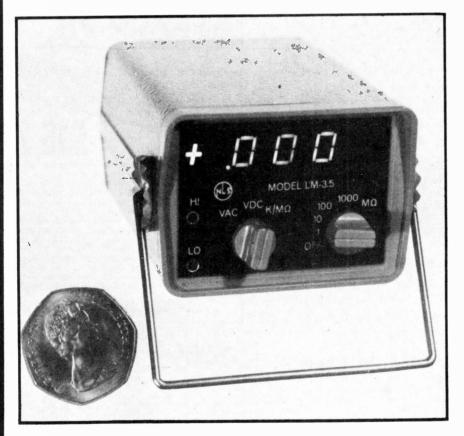
You can set 4 separate alarms and the time is accurate to ±15 seconds per month (0° - 40°C). The 199 year calender compensates for varying month lengths and leap years.

Finally, in addition to normal mathematical calculations, the CQ1,

allows calculations involving hours, minutes, seconds and elapsed between any two dates (1901 - 2100) to be computed. Price £35.95 RRP.

Casio Electronics Co. Ltd., 28 Scrutton Street, London, EC2.

#### FOR MEASURING SMALL MULTIS?



Four miniature digital multimeters made by Non Linear Systems Inc., are now available from Lawtronics. Each DMM is powered by internal rechargeable batteries, giving 2-3 hours operating life and is supplied with a charger unit.

Although each unit uses the same miniature case (1.9 x 2.7 x 4ins.), the

3, 3½, and 4 digit displays are 0.3ins. high, and have 25 ranges. The LM3, 3 digit is accurate to 1% of reading and the LM4, 4 digit is 0.03% of reading.

Instruments are normally supplied with a tilt stand. Prices start from £82, including batteries and charger.

Lawtronics, 139 High Street, Edenbridge, Kent TN8 5AX.

#### **GIANTS LETTING CMOS DOWN**

Across the pond in the USA there is a nice little session of throat-cutting going on. Texas are really clanging prices down on a range of MSI chips, including counters, multiplexers and registers.

Meanwhile back at the ranch National are reacting to Motorola's slash in January with similar hackings themselves. RCA on the other hand, have quietly taken their levels down without telling anybody.

#### FED UP SCRATCHING AROUND?

Researchers at University College London, have designed a prototype unit to eliminate completely surface clicks (scratches) from records. Completely different in concept to the usual top-cut filter, this unit detects the scratch (by monitoring the difference signal between the L and R channels), edits it from the output, and replaces it with an extrapolated signal which is indistinguishable from the original music.

The circuitry consists of a highpass filter (200Hz) automatic level control

and amplitude detector.

Upon recognising a scratch signal, the detector turns on a 'cross-fade'. A 6ms. delay is placed into the line to give the circuit time to do this. Once it finds a scratch, it waits 3ms. before firing the cross fader such that it edits all the scratch signal.

Don't go rushing down to your local hi-fi store and demand a demo just yet, however. Commercial exploitation of this wonder will depand on the ready supply of an A-D converter, at a price such that the overall circuit can be marketed at less than Britain's G.N.P.

#### ZILOG SECONDED

Mostek are second sourcing the Z80 MPU from Zilog. The two companies are working in cahoots as regards MPUs to produce this family of chips. Distronic distribute them in Britain.

The n-channel Z80 family includes high performance CPU, programmable parallel controller, programmable I/O controller, versatile counter-timer and high-speed direct memory access controller.

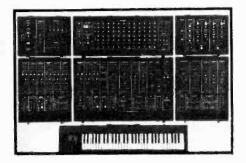
Distronic Ltd., 50/51 Burnt Mill, Elizabeth Way, Harlow, Essex.

#### **MURKOFILES SHIFTING**

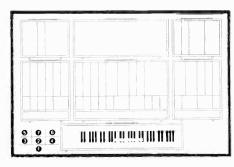
Signetics will be next with a single chip MPU, this to compete with Intel's 8048. It is an eight-bit system, will have 2K of ROM on board, and be titled 2645. Amen.

### THE WORLD'S BEST SYNTHESISER?

System 700 by Roland Electronics is claimed to be the ultimate in synthesisers. It is certainly quite a machine. The illustrated version costs £9000, although construction is modular. There just isn't enough room in the magazine to begin to describe



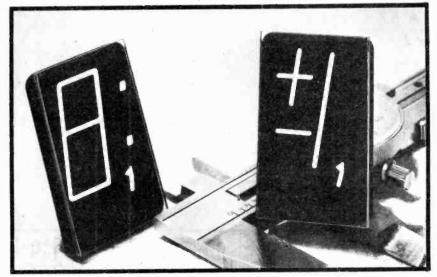
what it will do. One nice touch is the ability to be controlled by a musical instrument or voice. (There are 47 modules in the system.) Anyway if you're in London - go and hear it



yourself. It is on display at Freedmans, 629 High Road, Leytonstone, London E.11. Tell 'em ETI sent you!

P.S. If you know of a better synthesiser - let's hear about it!

#### **WHAT A GAS**



Some very clever displays just released from Beckman Instruments Ltd. are these SP101 and 102 GAS DIS-CHARGE packages! One inch high and looking remarkably like L.E.D.s, they require 160V DC at 70 uA to operate, being visible at 60 feet +, even in sunlight.

The colour is orange, and the characters are of the 'no gap between segment' type which can look very attractive. They are very slim, and well-suited to 'packed out' housing designs. Beckman Instruments Ltd., Queensway, Glenrothes, Fife, KY7 SPU.

#### THINK OF HIRE THINGS

It's worth remembering that when you need some particular piece of equipment - be it bionic or electronic - there is usually no need to mortgage the cat in order to buy it.

One very valid alternative exists, and that is to hire it. Most people need test gear for a limited time anyway, and purchase is not really economically justifiable in such cases.

A company called 'Livingstone Hire' will lease you whatever you need from a catalogue containing some 3,000 items. Most of their business is with companies big or small - but even if it's just little ole you, they'll probably be able to help.

Livingstone Hire Ltd., Shirley House, 27 Camden Road, London NW1 9NR.

## 5 WATT STEREO - THE CHIPS ARE DOWN!

We believe that some readers may be having difficulties in obtaining the version of the LM379 used in the 5 Watt Stereo project in our January

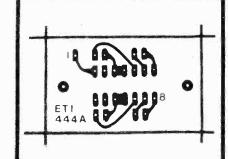
Maplin Electronics can supply readers with these devices, in the package which fits our PCB. The price is a reduced £4.56 all inclusive (from £5.43). For the address see the ad on the outside back cover. If you've already got the PCB, the new version of the LM379, won't fit - see the mod below.

## THE FASTEST DRAW(ER) IN THE WEST

Tektronix's new storage scope, the 7834, possesses a writing speed of 25000 mm/us., which is around 55,000,000 m.p.h., or about 1/9th. of the speed of light.

Single shot rise times as low as 1.4 ns., and repetitive signals down to 900 pico-seconds, can be displayed as the machine has a 400MHz mainframe. Plug-in modules extend the capabilities considerably.

Let's hope deliveries are as fast. Tektronix Ltd., Beaverton House, P.O. Box 69, Harpenden, Herts.



This small PCB will sit on top of the original board, enabling the new version of the power amplifier chip to be utilised simply in place of the package we used in the first place.

# mows dieost

#### **POSITIVELY REGULAR**

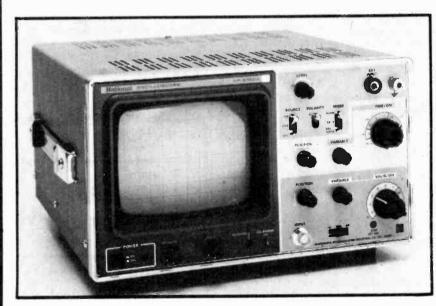


This three terminal, positive voltage regulator, capable of supplying in excess of 1.5 amps, and adjustable over an output range of 1.2V to 37V, has recently been announced by Jermyn. Requiring only two external resistors to set the output voltage, the LM317 has typical line and load regulation figures of 0.01%V and 0.1V acteristics to those of most standard

fixed voltage regulators. As this floating regulator sees only the input to output differential voltage, supplies of several hundred volts can be regulated providing the maximum differential of 40V is not exceeded. Current limit, thermal overload, and safe area protection are all incorporated within this regulator which also has 80dB ripple respectively, these being superior char-rejection and 1% temperature stability. Jermyn, Sevenoaks, Kent.

#### WATCH ORDERS AND ORDERS AND ORDERS AND...

We have had so many orders for our LCD watch offer that our original order was used up (twice over!) on the day of arrival. New stocks are available, and we are endevouring to keep any delay to a minimum. We hope to keep to our original 28-day order period, but if we don't make itplease be patient with us!



#### NATIONAL AT 10MHZ

This VP5100A is a general purpose DC-10MHz scope of 10mV maximum sensitivity. It is of reasonable proportions - 148 x 260 x 260mm. - and supposedly easy to lift about the place. Seven sweep rates from 0.1us. to 0.1s. per division are provided, as is TV mode for viewing video signals. It is being distributed by Telenic Altair, 2 Castle Hill Terrace, Maidenhead, Berks.

#### A CHIP OFF THE OLD EMPIRE!

A small British firm (aren't they all?) has taken on the might of the American semiconductor complexes in the battlefield of T.V. games. Sportel, as they are called, are producing their own COLOUR chip (30,000 a month in fact).

As well as all the usual party tricks, their 40 pin blob speeds up the ball after 4 bat strikes, automatically, and goes into 'Deuce type scoring in a game of tennis. Score is displayed as huge characters between one point being scored and the next service. The game will initially be offered as a builtin unit by Tyne marketing.

#### **RUSSIAN AROUND WITHOUT** PETROL

Soviet research has gotten a good way along the road to electric 'cars for the

people'.

One path - taken by Leningrad Technological Institute-leads to a 12kW air-magnesium storage battery, and employs a Moskvitch 408 as its base. Their leading contender however, is made in the Ukraine, and is based on a - wait for it - Zaporozhets vehicle which I hope is more driveable than it is pronounceable. Claimed range is 65 miles, and speed 45 m.p.h.

Why couldn't they call it a Mini or

something?

#### **RADAR RAIN**

Eighty towns in the USA use a computerised radar system to predict local weather. Oslo will be the first European convert to this meteorological microwave methodology.

The idea behind the system is to examine cloud formations and precipitation within a 200km. radius of the city. The equipment costs about £55,000 and does the work of 1,000

independent stations.

In this manner the computer provides accurate information of when and for how long rain or snow will fall. Relevant data will be stored on past weather, and all information presented on two V.D.U.s.

#### CAT OF THE YEAR?

We have just received our copy of the new Maplin Supplies catalogue. Only one word describes the publication superb! Looking rather like a telephone book in appearance, the book contains 216 pages on a very large page format (about the size of an ETI page) and has such a large range of components that Maplin have seen fit to include three indexes!

Six major projects are also included good range of books is also present. A superb job and well and truly worth the 50p you'll need to claim one.

# Metac Digital Clock

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We will invoice you with the clock. Try it out for 7 days then send your payment or return the clock in original condition.

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In choice of orange planar gas or soft green fluorescent digit displays. Green model has 24-hour readout. Orange model has 12-hour readout and AM/PM indicator. Both models have flashing second indicator, 24-hour bleeper alarm, 5-minute repeater, main failure indicator, 5" across x 3½" deep. Attractive white case. Thousands sold. Please state choice.

An electronic clock is silent and extremely reliable; because there are no moving parts it is impervious to dust or vibration and will continue to work indefinitely. Timing signals are derived from the 50 or 60Hz domestic electricity supply which in all the developed countries has to be held to very high levels of accuracy.

A bleeper alarm sounds until the clock is tipped forwards. The the "snooze" facility can give you 5 minutes sleep before the alarm sounds again, and then another 5 minutes, etc., until you switch the alarm off.

An indicator on the display tells you if the alarm is set, another indicator tells you if it's in the 'snooze' mode.

This remarkable clock even tells you if the electricity supply has momentarily failed.

**STOP PRESS** our **UXBRIDGE** shop is now open. Visit Metac-Electronics, Time Centre

3 NEW ARCADE, HIGH ST., UXBRIDGE, MIDDX.

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The Wireless Specialists for components & modules.

 $EF5800,\,7030~\&~91196$  -  $.9uV/30dB~S/N.,\,0.2\%~THD$  Our top three FM tunermodules.(EF5800 shown with can off).



From left to right, the EF5800 6 circuit varicap FM tunerhead with the 7030 linear phase IF and the 91196 PLL stereo decoder with integral 55kHz 'birdy' filter. The system provides afc muting, meter drives, agc, auto stereo switch, & a specification that exceeds broadcast requirements. Now available with a new EF5801 tunerhead, with FET buffered oscillator output for synthesiser/frequency readout facilities.

EF5801..£17.45; EF5800..£14.00; 7030..£10.95; 91196..£12.99

Complete FM tuner kits/systems (Carriage £3 extra.)

The Mark 8 Signalmaster - by Larsholt Electronics

This tuner is based on the popular 7252 tunerset, and provides an incomparable combination of style and performance that can be built by even the relatively inexperienced constructor. Complete kit...£85.00; matching 25+25W amplifier...£79.00.

International Mark 2 Tuner kits:

Complete tuner kit, based around the 7253 tunerset, £65.00. Or just the chassis, cabinet, heavey aluminium front panel — for your own choice of modules- see our new info. leaflet on the International Tuner. (SAE please)

# NEW NEW NEW NEW NEW NEW NEW BIONIC FERRET METAL LOCATOR

Ambit has designed a new approach to cost effective sensitive metal locators, and now we proudly present the first of the family of 'Bionic Ferrets'. Details OA, but we can say it will detect a 10p piece at 8-10 inches. Coupled with low power consumption and many innovations, this is the first radically advanced detector that can be made from a kit. £37.99

Radio module selection: (Prices for kits in our catalogue & PL)

EF5800	Ambit 6 stage varicap 88-108MHz tunerhead	£14.00
EF5600	TOKO 5 stage varicap ::	£12.95
EC3302	TOKO 3 stage varicap	£7.50
7020	Dual ceramic filter FM IF system module	£6.95
92310	MPX decoder, with stereo filter and preamp	£6.95
93090	MPX decoder with CA3090AQ + filter stage	£7.35
91197	MW/LW varicap AM tuner module	£11.35
771	New 'Off-Air' UHF varicap TV sound tunermoon	dule
l .	with mute, AFC, dual conversion, PSU	£27.00
9014	MW/LW/Stereo FM tuner chassis. Mech. tuned	£26.00

Components: ICs, coils, filters, trimmers diode law pots etc.

HA1137W/3089E FM IF 1.94
TBA120 FM IF and demod0.75
MC1350 FM IF preamp 0.97
SN76660N FM IF and det. 0.75
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TBA651 AM radio system 1.81
LA1197 AM radio system 1.81
LM197 AM radio system 1.81
LM380N 2 W Audio 1.05
LM381N audio preamp st. 1.81
TDA2020 20W audio amp 2.99
TCA940 10W audio amp 1.80
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# -ETI SUPPLEMENT-

R. M. Marsten describes thirty-five

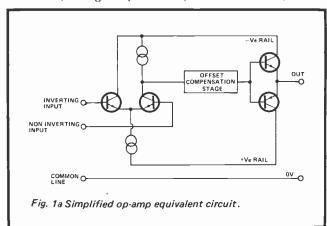
# 741 OP-AMP APPLICATIONS

OPERATIONAL AMPLIFIERS (OP-AMPS) CAN be simply described as high-gain direct-coupled voltage amplifier 'blocks' that have a single output terminal but have both inverting and non-inverting input terminals. Op-amps can readily be used as inverting, non-inverting, and differential amplifiers in both a.c. and d.c. applications, and can easily be made to act as oscillators, tone filters, and level switches, etc.

Op-amps are readily available in integrated circuit form, and as such act as one of the most versatile building blocks available in electronics today. One of the most popular, i.e. op-amps presently available is the device that is universally known as the '741' op-amp. In this article we shall describe the basic features of this device, and show a wide variety of practical circuits in which it can be used.

# BASIC OP-AMP CHARACTERISTICS AND CIRCUITS

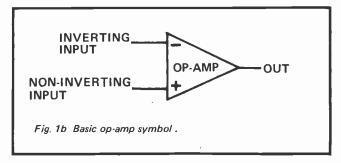
In its simplest form, an op-amp consists of a differential amplifier, followed by offset compensation and output stages, as shown in Fig. 1a. The differential amplifier has inverting and non-inverting input terminals, a high-impedance (constant current) tail to



give a high input impedance and a high degree of common mode signal rejection. It also has a high-impedance (constant current) load to give a high degree of signal voltage stage gain.

The output of the differential amplifier is fed to a direct-coupled offset compensation stage, which

effectively reduced the output offset voltage of the differential amplifier to zero volts under quiescent conditions, and the output of the compensation stage is fed to a simple complementary emitter follower output stage, which gives a low output impedance.



#### LINES OF SUPPLY

Op-amps are normally powered from split power supplies, providing +ve, -ve, and common (zero volt) supply rails, so that the output of the op-amp can swing either side of the zero volts value, and can be set at a true zero volts (when zero differential voltage is applied to the circuits input terminals.)

The input terminals can be used independently (with the unused terminal grounded) or simultaneously, enabling the device to function as an inverting, non-inverting, or differential amplifier. Since the device is direct-coupled throughout, it can be used to amplify either a.c. and d.c. input signals. Typically, they give basic low-frequency voltage gains of about 100 000 between input and output, and have input impedances of 1M or greater at each input terminal.

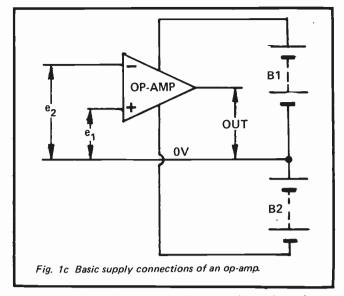
Fig. 1b shows the symbol that is commonly used to represent an op-amp, and 1c shows the basic supply connections that are used with the device. Note that both input and output signals of the op-amp are referenced to the ground or zero volt line.

#### **SIGNAL BOX**

The output signal voltage of the op-amp is proportional to the DIFFERENTIAL signal between its two input terminals, and is given by:

 $e_{out} = A_0(e_1 - e_2)$  where  $A_o =$  the open-loop voltage gain of the op-amp (typically 100 000).

ELECTRONICS TODAY INTERNATIONAL-APRIL 1977

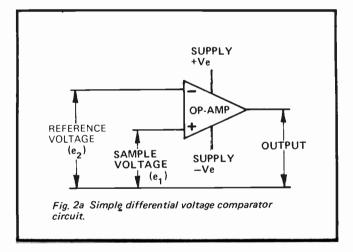


 $\boldsymbol{e}_1\!=\!\text{signal}$  voltage at the non-inverting input terminal.

e<sub>2</sub>=signal voltage at the inverting input terminal.

Thus, if identical signals are simultaneously applied to both input terminals, the circuit will (ideally) give zero signal output: If a signal is applied to the inverting terminal only, the circuit gives an amplified and inverted output: If a signal is applied to the non-inverting terminal only, the circuit gives an amplified but non-inverted output.

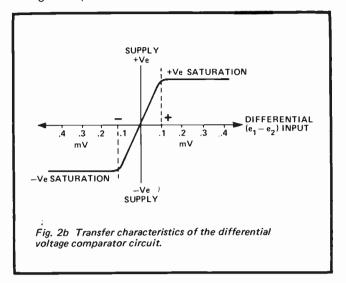
By using external negative feedback components, the stage gain of the op-amp circuit can be very precisely controlled.



#### TRANSFER REQUEST

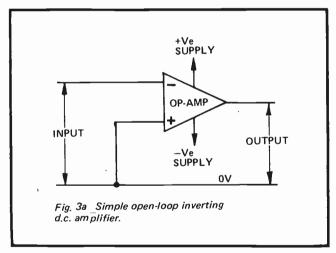
Fig. 2a shows a very simple application of the op-amp. This particular circuit is known as a differential voltage comparator, and has a fixed reference voltage applied to the inverting input terminal, and a variable test or sample voltage applied to the non-inverting terminal. When the sample voltage is more than a few hundred microvolts below the reference voltage the op-amp output is driven to saturation in a positive direction, and when the sample is more than a few hundred microvolts below the reference voltage the output is driven to saturation in the negative direction.

Fig. 2b shows the voltage transfer characteristics of the above circuit. Note that it is the magnitude of the differential input voltage that dictates the magnitude of the output voltage, and that the absolute values of input voltage are of little importance. Thus, if a 1V reference is used and a differential voltage of only 200uV is needed to switch the output from a negative to a positive saturation level, this change can be caused by a shift of only 0.02% on a 1V signal applied to the sample input. The circuit thus functions as a precision voltage comparator or balance detector.



#### GOING TO GROUND

The op-amp can be made to function as a low-level inverting d.c. amplifier by simply grounding the non-inverting terminal and feeding the input signal to

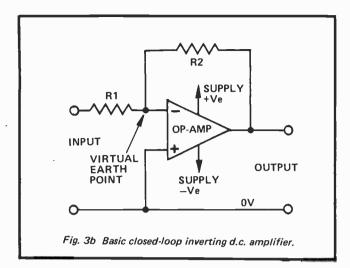


the inverting terminal, as shown in Fig. 3a. The op-amp is used 'open-loop' (without feedback) in this configuration, and thus gives a voltage gain of about 100 000 and has an input impedance of about 1M. The disadvantage of this circuit is that its parameters are dictated by the actual op-amp, and are subject to considerable variation between individual devices.

#### **CLOSING LOOPS**

A far more useful way of employing the op-amp is to use it in the closed-loop mode, i.e., with negative feedback. Fig. 3b shows the method of applying negative feedback to make a fixed-gain inverting d.c. amplifier. Here, the parameters of the circuit are controlled by feedback resistors  $R_1$  and  $R_2$ . The gain, A of the circuit is dictated by the ratios of  $R_1$  and  $R_2$ , and equals  $R_2/R_2$ .

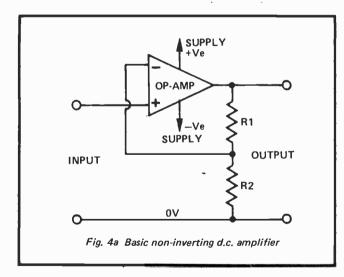
The gain is virtually independent of the op-amp characteristics, provided that the open-loop gain  $(A_o)$  is large relative to the closed-loop gain (A). The input impedance of the circuit is equal to  $R_1$ , and again is virtually independent of the op-amp characteristics.



#### **VIRTUALLY AT EARTH**

It should be noted at this point that although  $R_1$  and  $R_2$ 0 control the gain of the complete circuit, they have no effect on the parameters of the actual op-amp, and the full open-loop gain of the op-amp is still available between its inverting input terminal and the output. Similarly, the inverting terminal continues to have a very high input impedance, and negligible signal current flows into the inverting terminal. Consquently, virtually all of the  $R_1$  signal current also flows in  $R_2$ , and signal currents  $i_1$  and  $i_2$  can be regarded as being equal, as indicated in the diagram.

Since the signal voltage appearing at the output terminal end of  $R_2$  is A times greater than that appearing at the inverting terminal end, the current flowing in  $R_2$  is A times greater than that caused by the inverting terminal signal only. Consequently,  $R_2$  has an apparent value of  $R_2/A$  when looked at from its inverting terminal end, and the  $R_1-R_2$  junction thus appears as a low-impedance VIRTUAL EARTH point.



#### INVERT OR NOT TO INVERT . . .

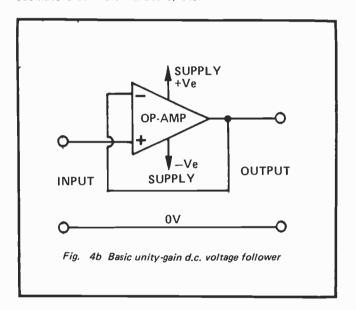
It can be seen from the above description that the Fig. 3b circuit is very versatile. Its gain and input impedance can be very precisely controlled by suitable choice of  $R_1$  and  $R_2$ , and are unaffected by variations in the op-amp characteristics. A similar thing is true of the non-inverting d.c. amplifier circuit shown in Fig. 4a. In this case the voltage gain is equal to  $(R_1+R_2)/R_2$  and the input impedance is approximately equal to

(A<sub>0</sub>/A)Zin where Zin is the open-loop input impedancw of the op-amp. A great advantage of this circuit is that it has a very high input impedance.

#### **FOLLOW THAT VOLTAGE**

The op-amp can be made to function as a precision voltage follower by connecting it as a unity-gain non-inverting d.c. amplifier, as shown in Fig. 4b. In this case the input and output voltages of the circuit are identical, but the input impedance is very high and is roughly equal to  $A_0 \times Z_m$ .

The basic op-amp circuits of Figs. 2a to 4b are shown as d.c. amplifiers, but can readily be adapted for a.c. use. Op-amps also have many applications other than as simple amplifiers. They can easily be made to function as precision phase splitters, as adders or subtractors, as active filters or selective amplifiers, as precision half-wave or full-wave rectifiers, and as oscillators or multivibrators, etc.



#### **OP-AMP PARAMETERS**

An ideal op-amp would have an infinite input impedance, zero output impedance, infinite gain and infinite bandwidth, and would give perfect tracking between input and output. Practical op-amps fall far short of this ideal, and have finite gain, bandwidth, etc., and give tracking errors between the input and output signals. Consequently, various performance paramaters are detailed on op-amp data sheets, and indicate the measure of 'goodness' of the particular device in question. The most importance of these parameters are detailed below.

**OPEN-LOOP VOLTAGE GAIN, A**<sub>o</sub>. This is the low-frequency voltage gain occuring directly between the input and output terminals of the op-amp, and may be expressed in direct terms or in terms of dB. Typically, d.c. gain figures of modern op-amps are 100 000, or 100dB.

**INPUT IMPEDANCE, Z**<sub>in</sub>. This is the impedance looking directly into the input terminals of the op-amp when it is used open-loop, and is usually expressed in terms of resistance only. Values of 1M are typical of modern op-amps with bi-polar input stages, while F.E.T. input types have impedances of a million Meg or greater.

**OUTPUT IMPEDANCE, Z\_0.** This is the output impedance of the basic op-amp when it is used open-loop, and is usually expressed in terms of resistance only. Values of a few hundred ohms are typical of modern op-amps.

**INPUT BIAS CURRENT, I\_b.** Many op-amps use bipolar transistor input stages, and draw a small bias current from the input terminals. The magnitude of this current is denoted by  $I_b$ , and is typically only a fraction of a microamp.

**SUPPLY VOLTAGE RANGE, V**<sub>s.</sub> Op-amps are usually operated from two sets of supply rails, and these supplies must be within maximum and minimum limits. If the supply voltages are too high the op-amp may be damaged, and if the supply voltages are too low the op-amp will not function correctly. Typical supply limits are  $\pm 3V$  to  $\pm 15V$ .

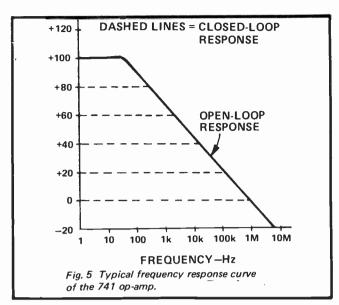
**INPUT VOLTAGE RANGE, V**<sub>i(max)</sub> The input voltage to the op-amp must never be allowed to exceed the supply line voltages, or the op-amp may be damaged.  $V_{i(max)}$  is usually specified as being one or two volts less than  $v_s$ .

**OUTPUT VOLTAGE RANGE, V**<sub>o(max)</sub>. If the opamp is over driven its output will saturate and be limited by the available supply voltages, so  $V_{o(max)}$  is usually specified as being one or two volts less than  $V_s$ .

**DIFFERENTIAL INPUT OFFSET VOLTAGE,**V<sub>ia.</sub> In the ideal op-amp perfect tracking would exist between the input and output terminals of the device, and the output would register zero when both inputs were grounded. Actual op-amps are not perfect devices, however, and in practice slight imbalances exist within their input circuitry and effectively cause a small offset or bias potential to be applied to the input terminals of the op-amp. Typically, this DIFFERENTIAL INPUT OFFSET VOLTAGE has a value of only a few millivolts, but when this voltage is amplified by the gain of the circuit in which the op-amp is used it may be sufficient to drive the op-amp output to saturation. Because of this, most op-amps have some facility for externally nulling out the offset voltage.

**COMMON MODE REJECTION RATION, c.m.r.r.** The ideal op-amp produces an output that is proportional to the difference between the two signals applied to its input terminals, and produces zero output when identical signals are applied to both inputs simultaneously, i.e., in common mode. In practical op-amps, common mode signals do not entirely cancel out, and produce a small signal at the op-amps output terminal. The ability of the op-amp to reject common mode signals is usually expressed in terms of common mode rejection ratio, which is the ratio of the op-amps gain with differential signals to the op-amps gain with common mode signals. C.m.r.r. values of 90dB are typical of modern op-amps.

**TRANSITION FREQUENCY,**  $f_T$  An op-amp typically gives a low-frequency voltage gain of about 100dB, and in the interest of stability its open-loop frequency response is tailored so that the gain falls off as the frequency rises, and falls to unity at a transition frequency denoted  $f_T$ . Usually, the response falls off at a rate of 6dB per octave or 20dB per decade. Fig. 5



shows the typical response curve of the type 741 op-amp, which has an  $f_{\rm T}$  of 1MHz and a low frequency gain of 100dB.

Note that, when the op-amp is used in a closed-loop amplifier circuit, the bandwidth of the circuit depends on the closed-loop gain. If the amplifier is used to give a gain of 60dB its bandwidth is only 1kHz, and if it is used to give a gain of 20dB its bandwidth is 100kHz. The  $f_{\mathsf{T}}$  figure can thus be used to represent a gain-bandwidth product.

	PARAMETER	741 VALUE
A <sub>0</sub>	OPEN-LOOP VOLTAGE GAIN	100dB
Z <sub>IN</sub>	INPUT IMPEDANCE	1M
z,	OUTPUT IMPEDANCE	150R
ı <sub>ь</sub>	INPUT BIAS CURRENT	200nA
V <sub>s</sub> (MAX)	MAXIMUM SUPPLY VOLTAGE	±18V
Vi (MAX)	MAXIMUM INPUT VOLTAGE	±13V
V <sub>D</sub> (MAX)	MAXIMUM OUTPUT VOLTAGE	±14V
Vio	DIFFERENTIAL INPUT OFFSET VOLTAGE	2mV
c.m.m.r.	COMMON MODE REJECTION RATIO	90dB
FT	TRANSITION FREQUENCY	1MHZ
s ·	SLEW RATE	1V/uS

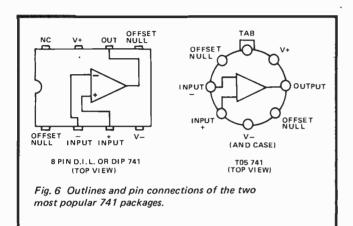
Table 1 Typical characteristics of the 741 op-amp.

**SLEW RATE.** As well as being subject to normal bandwidth limitations, op-amps are also subject to a phenomenon known as slew rate limiting, which has the effect of limiting the maximum rate of change of voltage at the output of the device. Slew rate is normally specified in terms of volts per microsecond, and values in the range 1V/uS to 10V/us are common with most popular types of op-amp. One effect of slew rate limiting is to make a greater bandwidth available to small output signals than is available to large output signals.

#### THE 741 OP-AMP.

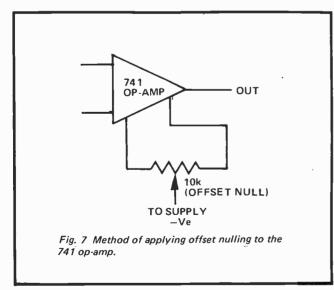
Early types of i.c. op-amp, such as the well known 709 type, suffered from a number of design weaknesses. In particular, they were prone to a phenomenon known as INPUT LATCH-UP, in which

the input circuitry tended to switch into a locked state if special precautions wre not taken when connecting the input signals to the input terminals, and tended to self-destruct if a short circuit were inadvertently placed across the op-amp output terminals. In addition, the op-amps were prone to bursting into unwanted oscillations when used in the linear amplifier mode, and required the use of external frequency compensation components for stability control.



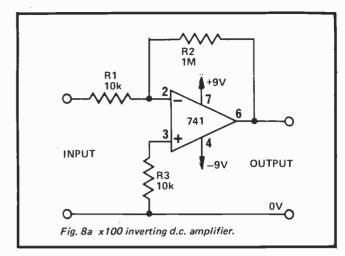
These weaknesses have been eliminated in the type 741 op-amp. This device is immune to input latch-up problems, has built-in output short circuit protection, and does not require the use of external frequency compensation components. The typical performance characteristics of the device are listed in Table 1.

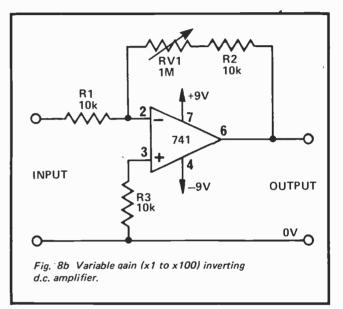
The type 741 op-amp is marketed by most i.c. manufacturers, and is very readily available. Fig. 6 shows the two most commonly used forms of packaging of the device. Throughout this chapter, all practical circuits are based on the standard 8-pin dual-in-line (D.I.L. or DIP) version of the 741 op-amp.



The 741 op-amp can be provided with external offset nulling by wiring a 10k pot between its two null terminals and taking the pot slider to the negative supply rail, as shown in Fig. 7.

Having cleared up these basic points, let's now go on and look at a range of practical applications of the 741 op-amp.





# BASIC LINEAR AMPLIFIER PROJECTS. (Figs. 8 to 11).

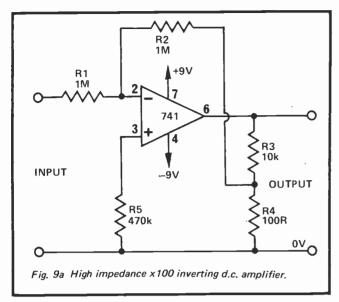
Figs. 8 to 11 show a variety of ways of using the 741 in basic linear amplifier applications.

The 741 can be made to function as an inverting amplifier by grounding the non-inverting input terminal and feeding the input signal to the inverting terminal. The voltage gain of the circuit can be precisely controlled by selecting suitable values of external feedback resistance. Fig. 8a shows the practical connections of an inverting d.c. amplifier with a pre-set gain of x100. The voltage gain is determined by the ratios of R<sub>1</sub> and R<sub>2</sub>, as shown in the diagram.

The gain can be readily altered by using alternative  $R_1$  and/or  $R_2$  values. If required, the gain can be made variable by using a series combination of a fixed and a variable resistor in place of  $R_2$ , as shown in the circuit of Fig. 8b, in which the gain can be varied over the range x1 to x100 via  $R_2$ .

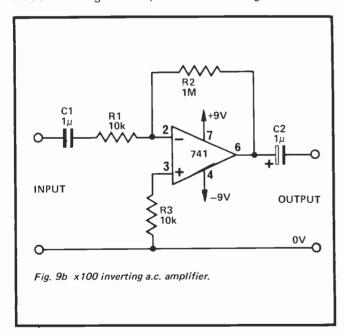
#### **VARIATIONS**

A variation of the basic inverting d.c. amplifier is shown in Fig. 9a. Here, the feedback connection to  $R_2$  is taken from the output of the  $R_3$ - $R_4$  output potential divider, rather than directly from the output of the op-amp, and the voltage gain is determined by the ratios of this divider as well as by the values of  $R_1$  and



R<sub>2</sub>. The important feature of this circuit is that it enables R<sub>1</sub>, which determines the input impedance of the circuit, to be given a high value if required, while at the same time enabling high voltage gain to be achieved.

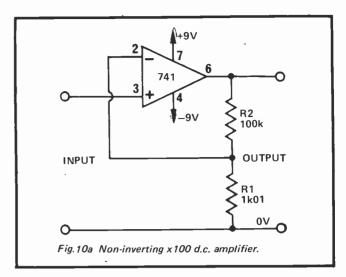
The basic inverting d.c. amplifier can be adapted for a.c. use by simply wiring blocking capacitors in series with its input and output terminals, as shown in the x100 inverting a.c. amplifier circuit of Fig. 9b.



#### NON-INVERTING . . .

The amp can be made to function as a non-inverting amplifier by feeding the input signal to its non-inverting terminal and applying negative feedback to the inverting terminal via a resistive potential divider that is connected across the op-amp output. Fig. 10a shows the connections for making a fixed gain (x100) d.c. amplifier.

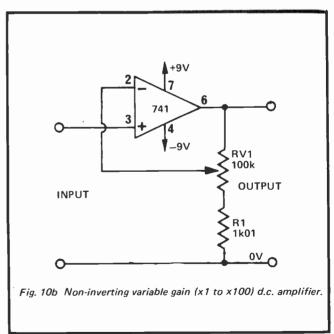
The voltage gain of the Fig. 10a circuit is determined by the ratios of  $R_1$  and  $R_2$ . If  $R_2$  is given a value of zero the gain falls to unity, and if  $R_1$  is given a value of zero the gain rises towards infinity (but in practice is limited to the open-loop gain of the op-amp). If required, the gain can be made variable by replacing  $R_2$  with a



potentiometer and connecting the pot slider to the inverting terminal of the op-amp, as shown in the circuit of Fig. 10b. The gain of this circuit can be varied over the range x1 to x100 via  $R_1$ .

#### ... AND RESISTANCE TO INPUTS

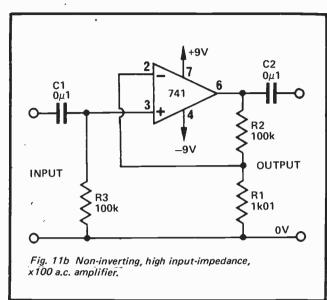
A major advantage of the non-inverting d.c. amplifier is that it has a very high input resistance. In theory, the input resistance is equal to the open-loop input resistance (typically 1M) multiplied by the open-loop voltage gain (typically 100 000) divided by the actual circuit voltage gain. In practice, input resistance values of hundreds of megohms can readily be obtained.



#### **BLOCKING OUT**

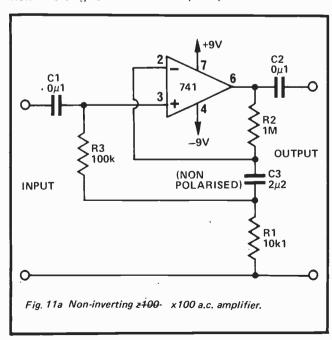
The basic non-inverting d.c. circuit of Fig. 10 can be modified to operate as a.c. amplifiers in a variety of ways. The most obvious approach here is to simply wire blocking capacitors in series with the inputs and outputs, but in such cases the input terminal must be d.c. grounded via a suitable resistor, as shown by R<sub>3</sub> in the non-inverting x100 a.c. amplifier of Fig. 11a. If this resistor is not used the op-amp will have no d.c. stability, and its output will rapidly drift into saturation.

Clearly, the input resistance of the Fig. 11a circuit is equal to  $R_3$ , and  $R_3$  must have a relatively low value in the interest of d.c. stability. This circuit thus loses the non-inverting amplifier's basic advantage of high input resistance.



#### **DRIFTING INTO STABILITY**

A useful development of the Fig. 11a circuit is shown in Fig. 11b. Here, the values of  $R_1$  and  $R_2$  are increased and a blocking capacitor is interposed between them. At practical operating frequencies this capacitor has a negligible impedance, so the voltage gain is still determined by the ratios of the two resistors. Because of the inclusion of the blocking capacitor, however, the inverting terminal of the op-amp is subjected to virtually 100% d.c. negative feedback from the output terminal of the op-amp, and the circuit thus has excellent d.c. stability. The low end of  $R_3$  is connected to the  $C_3$ – $R_1$  junction, rather than directly to the ground line, and the signal voltage appearing at this point is virtually identical with that appearing at the non-inverting terminal of the op-amp.



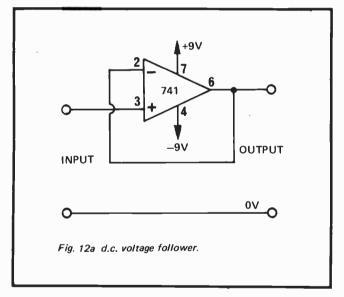
Consequently, identical signal voltages appear at both ends of R<sub>3</sub>, and the apparent impedance of this resistor is increased close to infinity by bootstrap action.

This circuit thus has good d.c. stability and a very high input impedance. In practice, this circuit gives a typical input impedance of about 50M.

#### **VOLTAGE FOLLOWER PROJECTS (Figs. 12 to 13).**

A 741 can be made to function as a precision voltage follower by connecting it as a unity-gain non-inverting amplifier. Fig. 12a shows the practical connections for making a d.c. voltage follower. Here, the input signal is applied directly to the non-inverting terminal of the op-amp, and the inverting terminal is connected directly to the output, so the circuit has 100% d.c. negative feedback and acts as a unity-gain non-inverting d.c. amplifier.

The output signal voltage of the circuit is virtually identical to that of the input, so the output is said to 'follow' the input voltage. The great advantage of this circuit is that it has a very high input impedance (as high as hundreds of megohms) and a very low output impedance (as low as a few ohms). The circuit acts effectively as an impedance transformer.



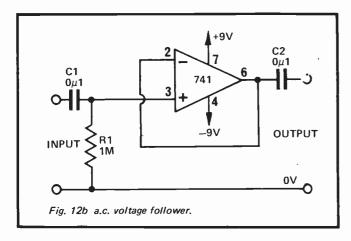
#### PRACTICE, AND ITS LIMITS

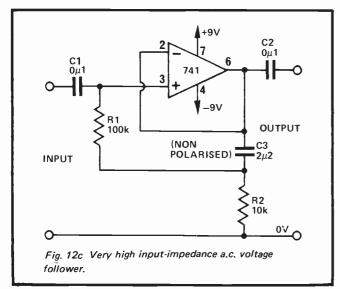
In practice the output of the basic Fig. 12a circuit will follow the input to within a couple of millivolts up to magnitudes within a volt or so of the supply line potentials. If required, the circuit can be made to follow to within a few microvolts by adding the offset null facility to the op-amp.

The d.c. voltage follower can be adapted for a.c. use by wiring blocking capacitors in series with its input and output terminals and by d.c.-coupling the non-inverting terminal of the op-amp to the zero volts line via a suitable resistor, as shown by  $R_1$  in Fig. 12b.  $R_1$  should have a value less than a couple of megohms, and restricts the available input impedance of the voltage follower.

#### **LACED UP OHMS**

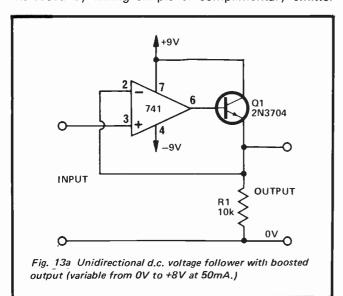
If a very high input-impendance a.c. voltage follower is needed, the circuit of Fig. 12c can be used. Here, R<sub>1</sub> is boostrapped from the output of the op-amp, and its apparent impedance is greatly increased. This circuit has a typical impedance of hundreds of megohms.





#### **DRIVING CIRCUITS AMP-LY**

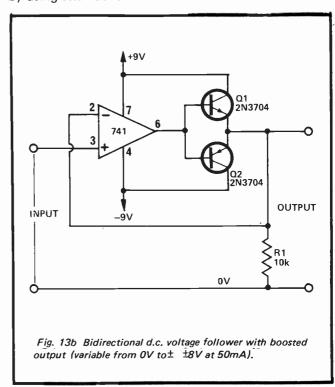
The 741 op-amp is capable of providing output currents up to about 5mA, and this is consequently the current-driving limit of the three voltage follower circuits that we have looked at so far. The current-driving capabilities of the circuits can readily be increased by wiring simple or complimentary emitter



follower booster stages between the op-amp output terminals and the outputs of the actual circuits, as shown in Figs. 13a and 13b respectively.

Note in each case that the base-emitter junction(s) of the output transistor(s) are included in the negative feedback loop of the circuit. Consequently, the 600mV knee voltage of each junction is effectively reduced by a factor equal to the open-loop gain of the op-amp, so the junctions do not adversely effect the voltage-following characteristics of either circuit.

The Fig. 13a circuit is able to source current only, and can be regarded as a unidirectional, positive-going, d.c. voltage follower. The Fig. 13b circuit can both source and sink output currents, and thus gives bidirectional follower action. Each circuit has a current-driving capacity of about 50mA. This figure is dictated by the limited power rating of the specified output transistors. The drive capability can be increased by using alternative transistors.



#### MISC AMP PROJECTS (Figs. 14 to 22)

Figs. 14 to 22 show a miscellaneous assortment of 741 amplifier projects, ranging from d.c. adding circuits to frequency-selective amplifiers.

Fig. 14 shows the circuit of a unity-gain inverting d.c. adder, which gives an output voltage that is equal to the sum of the three input voltages. Here, input resistors R<sub>1</sub> to R<sub>3</sub> and feedback resistor R<sub>4</sub> each have the same value, and the circuit thus acts as a unity-gain inverting d.c. amplifier between each input terminal and the output. Since the current flowing in each input resistor also flows in feedback resistor R4, the total current flowing in R<sub>4</sub> is equal to the sum of the input currents, and the output voltage is equal to the sum of the imput voltages. The circuit shown with only three input connections, but in fact can be provided with any number of input terminals. The circuit can be made to function as a so-called 'audio mixer' by wiring blocking capacitors in series with each input terminal and with the output terminal.

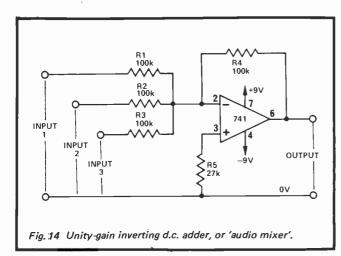
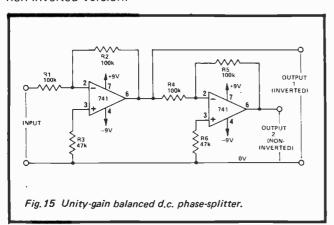


FIG. 15 shows how two unity-gain inverting d.c. amplifiers can be wired in series to make a precision unity-gain balanced d.c. phase-splitter. The output of the first amplifier is an inverted version of the input signal, and the output of the second amplifier is a non-inverted version.



**FIG. 16** shows how a 741 can be used as a unity-gain differential d.c. amplifier. The output of this circuit is equal to the difference between the two input signals or voltages, or to  $e_1$ - $e_2$ . Thus, the circuit can also be used as a subtractor. In this type of circuit the component values are chosen such that  $R_1/R_2=R_4/R_3$ , in which case the voltage gain  $A_v=R_2/R_1$ . The circuit can thus be made to give voltage gain if required.

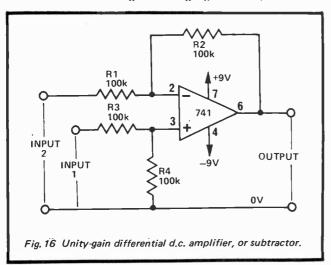
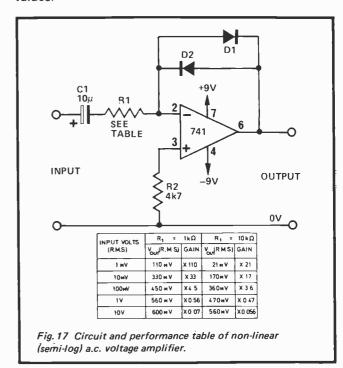


FIG. 17 shows the amp can be made to act as a non-linear (semi-log) a.c. voltage amplifier by using a couple of ordinary silicon diodes as feedback elements. The voltage gain of the circuit depends on the magnitude of applied input signal, and is high when input signals are low, and low when input signals are high. The measured performance of the circuit is shown in the table, and can be varied by using alternative R<sub>1</sub> values



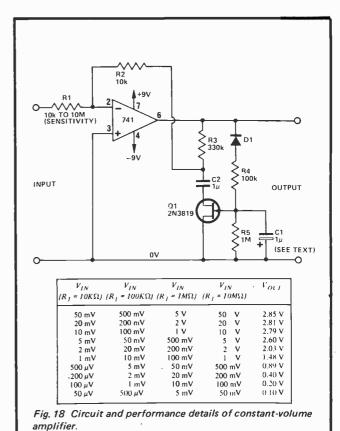


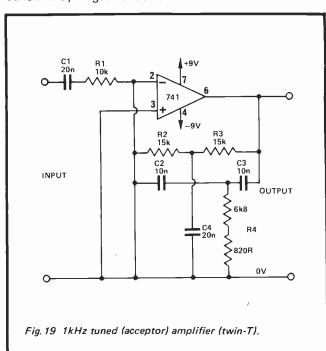
FIG. 18 shows how the 741 can be used together with a junction-type field-effect transistor (JFET) to make a so-called constant-volume amplifier. The action of this type of circuit is such that its peak output voltage is held sensibly constant, without distortion, over a wide range of input signal levels, and this particular circuit gives a sensibly constant output over a 30dB range of input signal levels.

The measured performance of the circuit is shown in the table.  $C_1$  determines the response time of the amplifier, and may be altered to satisfy individual needs.

#### **ACTION TAKEN**

The action of the Fig. 18 circuit relies on the fact that the JFET can act as a voltage-controlled resistance which appears as a low value when zero bias is applied to its gate and as a high resistance when its gate is negatively biased. The JFET and R<sub>3</sub> act as a gain-determining a.c. voltage divider (via C<sub>2</sub>), and the bias to the JFET gate is derived from the circuits output via the D<sub>1</sub>-C<sub>1</sub> network. When the circuit output is low the JFET appears as a low resistance, and the op-amp gives high voltage gain.

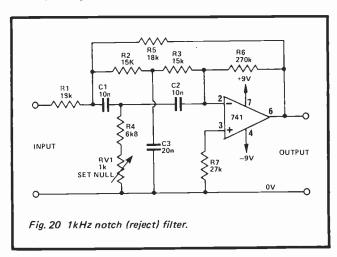
When the circuit output is high the JFET appears as a high resistance, and the op-amp gives low voltage gain. The output level of the circuit is thus held sensibly constant by negative feedback.

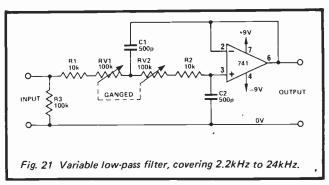


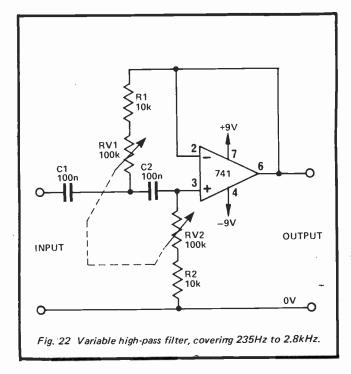
#### **CHOOSE YOUR FREQUENCY**

The 741 op-amp can be made to function as a frequency-selective amplifier by connecting frequency-sensitive networks into its feedback loops. Fig. 19 shows how a twin-T network can be connected to the op-amp so that it acts as a tuned (acceptor) amplifier, and Fig. 20 shows how the same twin-T network can be connected so that the op-amp acts as a notch (rejector) filter. The values of the twin-T network are chosen such that  $R_2 = R_3 = 2 \times R_4$ , and  $C_2 = C_4/2$ , in which case its centre (tuned) frequency =1/6.28  $R_2$ . $C_2$ . With the component values shown, both circuits are tuned to approximately 1kHz.

Finally, to complete this section, Figs. 21 and 22 show the circuits of a couple of variable-frequency audio filters. The Fig. 21 circuit is that of a low-pass filter which covers the range 2.2kHz to 24kHz, and the Fig. 22 circuit is that of a high-pass filter which covers the range 235Hz to 2.8kHz. In each case, the circuit gives unity gain to signals beyond its cut-off frequency, and gives a 2nd order response (a change of 12dB per octave) to signals within its range.

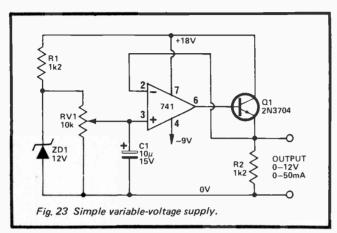






#### **INSTRUMENTATION PROJECTS (Figs. 23 to 31)**

Figs. 23 to 31 show a variety of instrumentation projects in which the 741 can be used. The circuits range from a simple voltage regulator to a linear-scale ohmmeter.

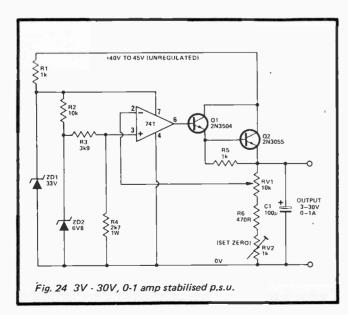


**FIG. 23** shows the circuit of a simple variable-voltage power supply, which gives a stable output that is fully adjustable from OV to 12V at currents up to a maximum of about 50mA. The operation of the circuit is quite simple. ZD<sub>1</sub> is a zener diode, and is energised from the positive supply line via R<sub>1</sub>. A constant reference potential of 12V is developed across the zener diode, and is fed to variable potential divider RV<sub>1</sub>.

The output of this divider is fully variable from 0V to 12V, and is fed to the non-inverting input of the op-amp. The op-amp is wired as a unity-gain voltage follower, with  $Q_1$  connected as an emitter follower current-booster stage in series with its output.

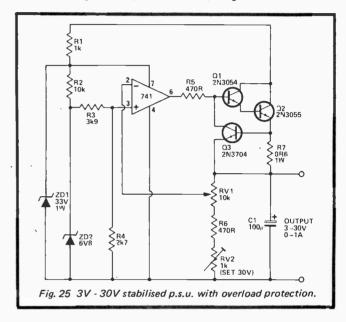
This, the output voltage of the circuit follows the voltage set at the op-amp input via  $RV_1$ , and is fully variable from 0V to 12V. Note that the circuit uses an 18V positive supply and a 9V negative supply.

Also note that the voltage range of the above circuit can be increased by using higher zener and unregulated supply voltages, and that its current capacity can be increased by using one or more power transistors in place of  $Q_1$ .



**FIG. 24** shows how a 741 op-amp can be used as the basis of a stabilised power supply unit (P.S.U.) that covers the range 3V to 30V at currents up to 1A. Here, the voltage supply to the op-amp is stabilized at 33V via  $ZD_1$ , and a highly temperature-stable reference of 3V is fed to the input of the op-amp via  $ZD_2$ .

The op-amp and output transistors  $Q_1$ - $Q_2$  are wired as a variable-gain non-inverting d.c. amplifier, with gain variable from unity to x10 via RV<sub>1</sub>, and the output voltage is thus fully variable from 3V to 30V via RV<sub>1</sub>. The output voltage is fully stabilized by negative feedback.



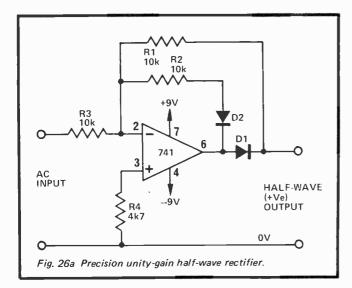
**FIG. 25** shows how overload protection can be applied to the above circuit. Here, current-sensing resistor  $R_9$  is wired in series with the output of the regulator, and cut-out transistor  $Q_3$  is driven from this resistor and is wired so that its base-collector junction is able to short the base-emitter junction of the  $Q_1$ - $Q_2$  output transistor stage.

Normally,  $\Omega_3$  is inoperative, and has no effect on the circuit, but when P.S.U. output currents exceed 1A a potential in excess of 600mV is developed across  $R_9$  and biases  $\Omega_3$  on, thus causing  $\Omega_3$  to shunt the base-emitter junction of the  $\Omega_1$ - $\Omega_2$  output stage and hence reducing the output current. Heavy negative feedback takes place in this action, and the output current is automatically limited to 1A, even under short-circuit conditions.

**FIG. 26a** shows how a 741 can be used in conjunction with a couple of silicon diodes as a precision half-wave rectifier. Conventional diodes act as imperfect rectifiers of low-level a.c. signals, because they do not begin to conduct significantly until the applied signal voltage exceeds a 'knee' value of about 600mV.

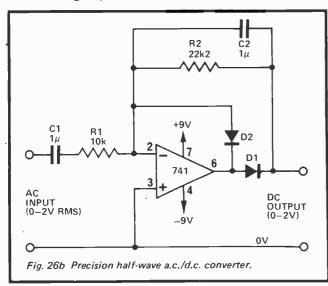
When diodes are wired into the negative feedback loop of the circuit as shown the 'knee' voltage is effectively reduced by a factor equal to the open-loop gain of the op-amp, and the circuit thus acts like a near-perfect rectifier.

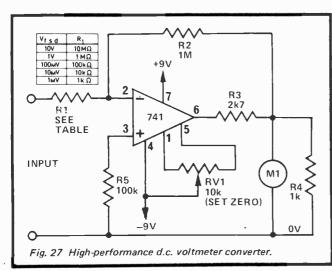
The overall voltage gain of the Fig. 26a circuit is dictated by the ratios of  $R_1$  and  $R_2$  to  $R_3$ , as in the case of a conventional inverting amplifier, and this circuit thus gives a gain of unity. The circuit can be made to



act as a precision half-wave a.c./d.c. converter by designing it to give a voltage gain of 2.22 to give form-factor correction, and by integrating its rectifier output, as shown in Fig. 26b.

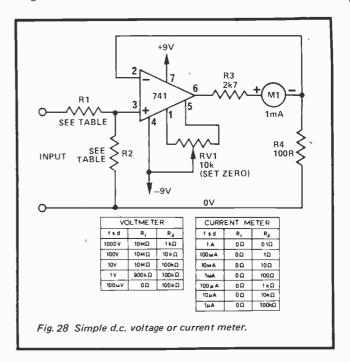
Note that each of the Fig. 26 circuits has a high output impedance, and the outputs must bot be fed into loads having impedances less than about 1M.





**FIG. 27** shows how op-amp can be used as a high-performance d.c. voltmeter converter, which can be used to convert any 1V f.s.d. meter with a sensitivity better than 1k/V into a voltmeter that can read any value in the range 1mV to 10V f.s.d. at a sensitivity of 1M/V. The voltage range is determined by the R<sub>1</sub> value, and the table shows some suitable values for common voltage ranges.

FIG. 28 shows a simple circuit that can be used to convert a 1mA f.s.d. meter into a d.c. voltmeter with any f.s.d. value in the range 100mV to 1000V, or into a d.c. current meter with any f.s.d. value in the range 1uA to 1A. Suitable component values for different ranges are shown in the tables.



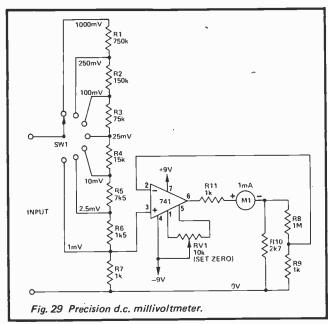
**FIG. 29** shows the circuit of a precision d.c. millivoltmeter, which uses a 1mA f.s.d. meter to read f.s.d. voltages from 1mV to 1000mV is seven switch-selected ranges.

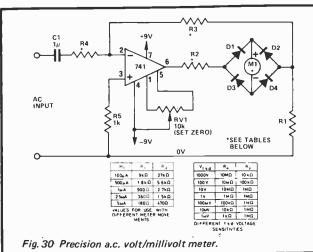
**FIG. 30** shows the basic circuit of a precision a.c. volt or millivolt meter. This circuit can be used with any moving-coil meter with a full scale current value in the range 100uA to 5mA, and can be made to give any full scale a.c. voltage reading in the range 1mV to 1000mV. The tables show the alternative values of  $R_1$  and  $R_2$  that must be used to satisfy different basic meter sensitivities, and the values of  $R_3$  and  $R_4$  that must be used for different f.s.d. voltage sensitivities.

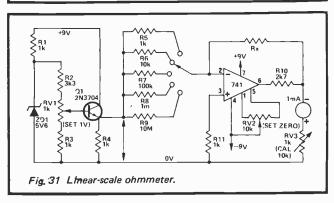
#### HOME OHM

Finally, to conclude, Fig. 31 shows how the 741 op-amp can be used in conjuncton with a 1mA f.s.d. meter to make a linear-scale ohmmeter that has five decade ranges from 1k to 10M.

The circuit is divided into two parts, and consists of a voltage generator that is used to generate a standard test







voltage, and a readout unit which indicates the value of the resistor under test.

The voltage generator section of the circuit comprises zener diode  $\mathrm{ZD}_1$ , transistor  $\mathrm{Q}_1$ , and resistors  $\mathrm{R}_1$  to  $\mathrm{R}_4$ . The action of these components is such that a stable reference potential of 1V is developed across  $\mathrm{R}_4$ , but is adjustable over a limited range via  $\mathrm{RV}_1$ . This voltage is fed to the input of the op-amp readout unit. The op-amp is wired as an inverting d.c. amplifier, with the 1mA meter and  $\mathrm{RV}_3$  forming a 1V f.s.d. meter across its output, and with the op-amp gain determined by the

values of ranging resistors  $R_{\rm 5}$  to  $R_{\rm 9}$  and by negative feedback resistor  $R_{\rm x}$ .

Since the input to the amplifier is fixed at 1V, the output voltage reading of the meter is directly proportional to the value of  $R_{\rm x}$ , and equals full scale when  $R_{\rm x}$  and the ranging resistor values are equal. Consequently, the circuit functions as a linear-scale ohmmeter.

#### **CALIBRATION**

The procedure for intially calibrating the Fig. 1.31 circuit is as follows. First, switch the unit to the  $10k\Omega$  range and fix an accurage  $10k\Omega$  resistor in the  $R_x$  position. Now adjust  $R_3$  to give an accurate 1V across  $R_5$ , and then adjust  $R_{12}$  to give a precise full scale reading on the meter. All adjustments are then complete, and the circuit is ready for use.

#### **MISCELLANEOUS 741 PROJECTS**

The 741 op-amp can be used as the basis of a vast range of miscellaneous projects, including osillators and sensing circuits. Four such projects are described in this final section.

**FIG. 32** shows how the 741 op-amp can be connected as a variable-frequency wien-bridge oscillator, which covers the basic range 150Hz to 1.5kHz, and uses a low-current lamp for amplitude stabilisation. The output amplitude of the oscillator is variable via RV<sub>4</sub> and has a typical maximum value of 2.5V r.m.s. and a t.h.d. value of 0.1%. The frequency range of the circuit is inversely proportional to the  $C_1$ - $C_2$  values. The circuit can give a useful performance up to a maximum frequency of about 25kHz.

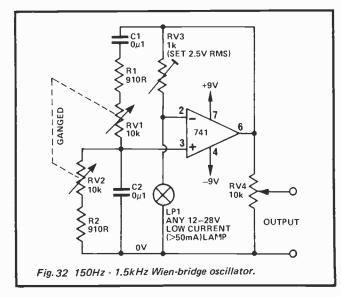
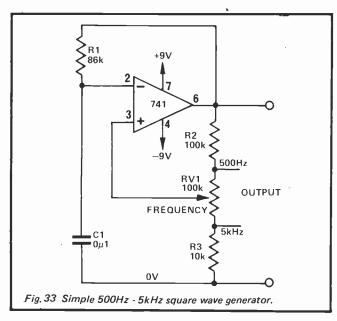


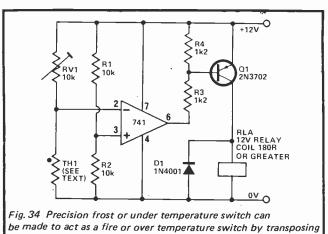
Fig. 33 shows how either a 741 or a 709 op-amp can be connected as a simple variable-frequency square-wave generator that covers the range 500Hz to 5kHz via a single variable resistor. (The circuit produces a good symmetrical waveform.)

The frequency of oscillation is inversely proportional to the C<sub>1</sub> value, and can be reduced by increasing the C<sub>1</sub> value, or vice-versa. The amplitude of the square wave output signal can be made variable, if required, by wiring a  $10k\Omega$  variable potential divider across the output terminals of the circuit and taking the output from between the pot slider and the zero volts line.



**FIGS. 34 and 35** show a couple of useful ways of using the 741 op-amp in the open-loop differential voltage comparator mode. In each case, the circuits are powered from single-ended 12V supplies, and have a fixed half-supply reference voltage applied to the non-inverting op-amp terminal via the  $R_1$ - $R_2$  potential divider and have a variable voltage applied to the inverting op-amp terminal via a variable potential divider.

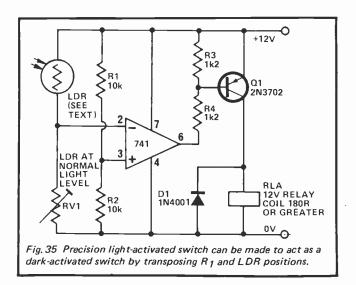
The circuit action is such that the op-amp output is driven to negative saturation and the relay is driven on when the variable input voltage is greater than the reference voltage, and the op-amp output is driven to positive saturation and the relay is cut off when the variable input voltage is less than the reference voltage.



#### FROSTY RECEPTION

R<sub>1</sub> and TH<sub>1</sub> positions.

The Fig. 34 circuit is that of a precision frost or under-temperature switch, which drives the relay on when the temperature sensed by thermistor TH $_1$  falls below a value pre-set via RV $_1$ . The circuit action can be reversed, so that it operates as a fire or over-temperature switch, by simply transposing the RV $_1$  and the TH, positions. In either case, TH $_1$  can be any negative-temperature-coefficient thermistor that presents a resistance in the range  $900\Omega$  to  $9k\Omega$  at the required trip temperature.



#### **LIGHT WORK**

The Fig. 35 circuit is that of a precision light-activated switch, which turns the relay on when the illumination level sensed by light-dependent resistor LDR exceeds a value pre-set by  $RV_1$ . The circuit action can be reversed so that the relay turns on when the illumination falls below a pre-set level by simply transposing the  $RV_1$  and LDR positions. In either case, the LDR can be any cadmium-sulphide photocell that presents a resistance in the range  $900\Omega$  to  $9k\Omega$  at the desired switch-on level.



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CATALOGUE 20p

CLOCK CHIPS MM5314 4.25 MM5316 5.95 MAY-5-1224A 3.25 MAY-5-4007D	D10DES BA145 BA148 BA155 BA156 BA157 BA158 BA159	0.14 0.13 0.13 0.12 0.22 0.22
C SOCKETS 3 pm 0.16 14 pm 0.16 16 pm 0.18 24 pm 0.45 10 pm 0.80	BY206 BY207 BYX36 -300 -600 -900 -1200 BYX38	0.15 0.20 0.12 0.15 0.18 0.21
	-300	0.50

<b>REGULAT</b> 723 7805 7812 7815	ORS 0.45 1.50 1.50 1.50
7815	
7818 LM340-5 LM340-12	1.35
LM340-15 LM340-18	
TOELECT	RONICS

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		(
PTOELECT	RONICS	(
Displays		(
704	0.99	(
707	0.99	1
727	1.95	1
728	1.95	1
747	1.80	1
750	1.80	- 1
LED		!
0.0.	0.12	: 1

	BYX38	
-	-300	0.50
		0.65
	-900	0.60
	-1200	
	BZX61	0.05
		0.20
	series	0.20
	BZX83	
	series	0.11
	BZY88	
	series	
	OA5	0.50*
	OA10	0.40
$\neg$	OA85	0.12
s	OA90	0.08
	OA91	0.08
	0A200	0.09
		0.04+
	IN4001	
	IN4002	
	IN4003	
	IN4004	
	IN4005	
	IN4005	
	IN4007	
	IN4148	U.U4*

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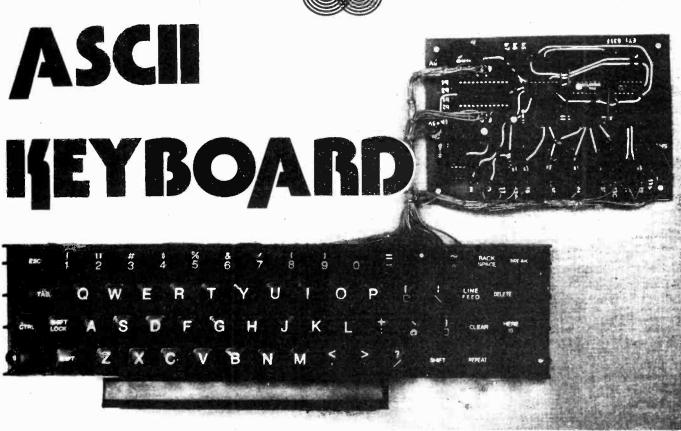
OCP71 £1.15 RECTIFIERS DO-4 PACKAGE

10A 50V 0.40 10A 100V 0.45 10A 200V 0.50 10A 400V 0.60 Please specily Polarity Stud Cathode or Stud Anode Ideal for Power Supplies, Inverters, etc

TRANS	ISTORS								
		BC171	0.12*	BD183	0.97	BFY64	0.35	2N1306	0.35
AC126	0.15	BC1728	0.12*	BD184	1.20	BFY90	0.65	2N1307	0.35
AC127	0.16	BC182	0.11*	BD232	0.60	BLY15A		2N1308	0.45
AC128	0.13	8C182L	0.12*	BD233	0.48	BSX19	0.16	2N1309	0.45
AC128K	0.25	BC183	0.10*	BD237	0.55	BSX20	0.18	2N1711	0.18
AC141	0.22	BC183L	0.10*	BD238	0.60	BSX21	0.20 0.30	2N2102 2N2217	0.44
AC141K	0.34	.BC184	0.11*	BD410	0.60 2.30	85x76	0.30	2N2217 2N2369	0.30
AC142	0.18	BC184L	0.12*	BDX32 BDY10	1.50	BSX77 BSX78	0.30	2N2369A	0.14
AC142K AC176	0.28 0.16	BC186 BC187	0.20* 0.24*	BDY 10	2.00	BSY52	0.36	2N2309A 2N2483	0.20
AC176 AC176K	0.16 0.25	BC187 BC207B	0.24*	BDY20	0.80	BSY53	0.39	2N2484	0.16
AC176K AC187	0.25	BC207B BC212	0.12* 0.11*	BDY38	0.60	8SY54	0.44	2N2646	0.50
AC187K	0.25	BC212L	0.12*	BDY60	1.70	BSY55	0.74	2N2711	0.20
AC188	0.18	BC213	0.12*	BDY61	1.65	BSY65	0.30	2N2712	0.15
AC188K	0.25	BC213L	0.14*	BDY62	1.15	BSY76	0.20	2N2904A	0.20
AD149	0.45	BC214	0.14±	BOY95	2.14	BSY78	0.75	2N2905	0.18
AD161	0.35	BC214L	0.15*	BF121	0.50	BSY95A	0.12	2N2905A	0.22
AD162	0.35	BC237	0.16*	BF123 BF127	0.50 0.50	BU105/02	1.80*	2N2906 2N2925	0.18 0.14*
AF114	0.20	BC238	0.16*	BF127 BF157	0.50	BU105/02 BU108	1.90* 3.00*	2N2925 2N29260	0.14*
AF115	0.20 0.20	BC300	0.34 0.32	BF157	0.25	BU 108	3.00* 2.50*	2N29260 2N2926R	0.10*
AF116 AF117	0.20	BC301 BC302	0.32	BF178	0.28	BU126	1.60*	2N2926Y	0.09*
AF118	0.20	BC302	0.46	BF179	0.30	BU133	1.60*	2N2926G	0.10*
AF124	0.25	BC307	0.15*	BF1BQ	0.30	BU204	1.60*	2N3053	0.15
AF125	0.25	BC308	0.16*	BF181	0.30	BU205	1.90*	2N3055	0.50
AF126	0.25	BC309	0.18*	BF182	0.30	BU206	2.40*	2N3133	0.30
AF139	0.35	BC310	0.20*	BF183	0.30	BU208	2.60*	2N3134	0.30 1.10
AF239	0.37	BC317	0.12*	BF184	0.20	MJ480	0.80	2N3137 2N3440	0.56
AL102	0.95	BC319	0.13* 0.18*	BF185 BF194	0.20	MJ481 MJ490	1.05 0.90	2N3440 2N3442	1.20
AL103	0.93 3.30+	BC320 BC321	0.18* 0.18*	BF194 BF196	0.10* 0.12*	MJ490 MJ491	1.15	2N3442 2N3570	0.80
AU107 AU110	3.30 * 1.75 *	BC321	0.60*	BF196	0.12*	MJE340	0.40*	2N3702	0.10*
AU110	1.75* 1.60*	BC323	0.18*	BF200	0.40	MJE520	0.45	2N3703	0.10*
BC107	0.09	BC328	0.16*	8F218	0.30	MJE521	0.55	2N3704	0.10*
BC107B	0.09	BC337	0.17*	BF219	0.30	OC43	0.95	2N3705	0.10*
BC108	0.09	BC338	0.17*.	BF220	0.28	OC44	0.32	2N3706	0.10*
BC108B	0.09	BC407	0.22*	BF224J	0.18*	OC45	0.32	2N3/07	0.10*
BC109 '	0.09	BC408	0.22*	BF244	0.17*	0046	0.20	2N3708 2N3709	0.09*
BC109B	0.09	BCY30 BCY31	0.55	BF257	0.30	OC70 OC71	0.30	2N3709 2N3710	0.09*
BC109C BC117	0.12 0.19*	BCY31 BCY32	0.55 0.60	BF258 BF259	0.35	OC72	0.22	2N3710 2N3711	0.10*
BC117 BC119	0.19*	BCY33	0.55	BF259 BF336	0.35*	OC84	0.40	2N3715	1.15
BC119	0.18+	8CY34	0.55	BF337	0.32*	OC139	1.30	2N3716	1.25
BC126	0.20*	BCY38	0.50	BF338	0.45*	OC140	1.30	2N3771	1.60
BC140	0.32	BCY39	1.15	BFW30	0.06	OC170	0.23	2N3772	1.60
BC141	0.28	BCY40	0.75	BFW59	0.16	TIP29A	0.44*	2N3773	2.10
BC142	0.23	BCY42	0.30	BFW60	0.17	TIP30A	0.52*	2N3819	0.28* 0.16*
BC143	0.23	BCY54 BCY70	1.60 0.12	BFX29 BFX30	0.26 0.30	TIP31A TIP32A	0.54 0.64	2N3904 2N3906	0.10*
BC144	0.30 0.09*	BCY70 BCY71	0.12	BFX84	0.30	TIP32A	0.68	2N3900 2N4123	0.14*
BC147 BC148	0.09*	BCY72	0.12	BFX85	0.25	TIP42A	0.72	2N4124	0.14*
BC148	0.09*	BD115	0.15	BFX86	0.25	2N404	0.40	2N4290	0.14*
BC152	0.25*	BD131	0.36	BFX87	0.20	2N696	0.14	2N4291	0.14*
BC153	0.18*	BD132	0.40	BFX88	0.20	2N697	0.12	2N4292	0.14*
BC157	0.09*	80135	0.36*	BFX89	0.90	2N706	0.10	2N4347	1.10
BC158	0.09*	BD136	0.39*	BFY11	1.10	2N718	0.22	2N4348	1.20 0.35*
BC159	0.09*	BD137 BD138	0.40 ± 0.48 ±	BFY18	0.50	2N929 2N1131	0.14	2N4870 2N4871	0.35 * 0.35 *
BC160 BC161	0.32 0.36	BD138	0.48*	BFY40 BFY41	0.50	2N1131 2N1132	0.16	2N4871 2N4918	0.60*
BC168	0.38	BD144	2.20	BFY50	0.20	2N1302	0.25	2N4919	0.70*
BC169	0.12*	BD157	0.80	BFY51	0.18	2N1303	0.27	2N4920	0.50*
BC169C	0.14*	BD181	0.86	BFY52	0.19	2N1304	0.45	2N4922	0.58*
BC170B	0.12*	BD182	0.92	BFY53	0.25	2N1305	0.24	2N4923	0.46*







THIS PROJECT, NUMBER 631, FORMS THE FIRST PART OF OUR MPU HOME SYSTEM, THE ENCODER BOARD PROVIDES FULL ASCII OUTPUT, AND CAN BE USED WITH MOST COMMERCIAL KEYBOARD UNITS.

TO COMMUNICATE WITH A computer you need some sort of input device and some sort of output reader. The input unit can be a series of switches on which you set up the required code and press a button to enter each character. While this is economical in parts it is not economical in time.

This encoder project is designed to allow very easy access to the computer whilst being reasonably economical. It is very flexible and allows for almost any keyboard to be used. Control functions can be activated by a single key if desired and lower case letters can be eliminated at the flick of a switch.

The output from the keyboard is in the form of a parallel bus and the data has to be serialised to provide a universal input which will then communicate with any computer designed to work with a teletype.

#### **DESIGN FEATURES**

When we first looked at a keyboard encoder we intended to

use a single chip device to simplify design. However, looking at the devices available and their limitations (and cost and availability) it was decided to compromise and use the HD0165 keyboard encoder. This IC has been available for many years and we use it to decode the first 4 lines. For the other three lines we decided to use discrete components. The eighth line is not used at this stage (it is used for a parity check after serialisation).

Initially the use of a 16 x 3 matrix was contemplated. Then we would use the shift and control keys to get the other outputs. However, not all keys with the same three-line code (b5, 6, 7) are upper case (or lower case). On our keyboard 0 1 ...; are lower case, and = ? are upper case; yet all have an output code 3. The same applies to other rows and the matrix has thus expanded to 16 x 7. To get the control functions a control and the function key have to be pressed simultaneously, which is inconvenient for commonly-used

functions (such as space or line feed).

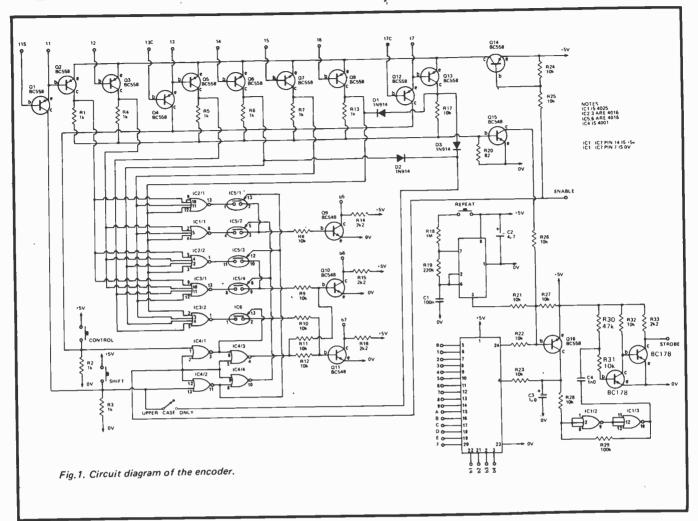
Consequently an additional three lines are used and this allows any of the control functions to be activated by a single key.

Most VDUs or microcomputer operating systems cannot handle lower-case letters and therefore outputs are provided which can be linked to ensure that a shift command is given automatically when any key from A-Z is pressed.

When connecting to the keyboard we had to decide how to wire the contacts. The easiest and neatest way is to use a double-sided pc board with plated-through holes. Using such a board it is hard to solder the other side when it is against the keyboard!

The alternative, and the approach we chose, is to link the underside of the keyboard using "solderable" enamelled wire and normal hookup wire to the control card. This takes a little time to wire but is much cheaper and is universal. Although we used a

# -SYSTEM 6S



# How it works

The Harris HDO165 IC is a 16-line keyboard encoder; if any one of its 16 inputs is taken high (+5V) an output code appears on the four output lines. At the same time another output (pin 4) goes low and another output (pin 24) goes low to indicate this.

In this project we use this IC to generate the least significant four bits (b1, b2, b3, and b4) of the seven bits we need to represent the complete character.

To decode the other three bits we used discrete transistors and CMOS gates. Each key joins one of the inputs of the HDO165 to one of the points 1-17. If the enable line is low (i.e. 0V) Q14 will be hard on and we will have 5V (less a little) on the emitter of Q2,3,5,6,7,8 and 13. The input of the HDO165 appears as a resistor of about 500-600 ohms, to 0V. Therefore connecting (say) point 14 to point 3, we turn on Q6 giving +5V at its collector and also the HDO165 gives an output corresponding to three (0011).

The high output from Q6 gives a high on the inputs of IC 2/2 and IC3/2 causing the outputs of these gates to be low. The other gates, IC1/1; IC2/1 and IC3/1

have high outputs. If the control or shift key is not pressed, we have a '0' at the input of IC4/1 and IC4/2 giving a high output from these gates and hence a low output from IC4/3 and IC4/4. This enables IC5/1, IC5/3 and IC6. These ICs are simply electronic switches with a resistance of either 300 ohms (on) or infinity (off).

Therefore Q9 will be on as IC2/1 is high, Q10 will be off as IC2/2 is low and Q12 will also be off as IC3/2 is low. This gives a total output of  $110\ 0011$  which represents 63 (hex) or lower case c.

We will leave you to work out the other combinations. If the shift key is pressed, IC5/2, IC5/4 and IC6 are enabled selecting a different code (upper case C is 43 hex) and if the control key is pressed, Q10 and 11 are turned on by IC4/3 and Q9 is controlled by IC1/1 and IC2/1 ('control C' is 03 (hex), representing ETX).

When a key is pressed the output (pin 4) of the HDO165 goes low and C3 is discharged via R23. After about 10ms. the gates IC1/2, 3, which are connected as a schmitt trigger, operate and the out-

put (1C1/3) goes low. This is coupled via C4. Q17/18 act as a monostable giving a negative-going pulse of about 200µs wide. When one key is pressed about 0.4V is developed across R20, not quite enough to turn on Q15. If a second key is pressed in a different row, the additional current in R20 will forward-bias Q15 which will then turn on Q16. This holds C3 charged, independent of the HDO165. If two keys are pressed in the same row on output (pin 24), the HDO165 detects this and goes low and Q16 is again turned on disabling the strobe pulse.

If the repeat button is pressed IC7 oscillates at about 10Hz and the pulsing alternately turns Q16 on and off generating strobe pulses at about 10 per sec.

The output of Q6 (A-O) and Q17 (P-Z) are diode ORed and if the 'upper case only' link or switch is closed it automatically gives a shift command. For the control functions additional inputs are used in Q1, Q4 and Q12. If the input to one of these transistors is connected to one of the HDO165 inputs it still turns on the transistor associated with it and also lifts either the control or shift inputs as required.



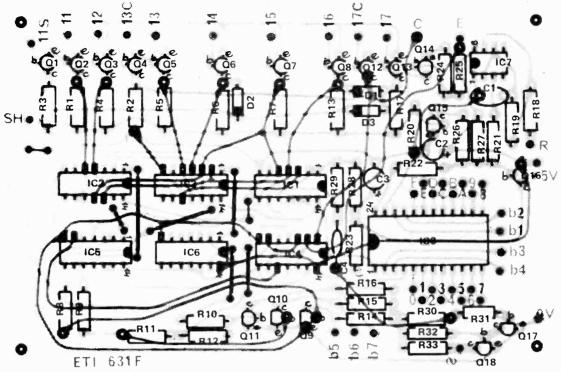


Fig.2. Component overlay of the encoder.

double-sided board for the control logic we don't require plated-through holes, as both sides can be easily soldered.

#### CONSTRUCTION

Assemble the PCB board with the aid of the overlay in Fig 2. When soldering the components use a small iron and make sure all connections on the component side are soldered as well as those on the copper side. The links on the component side must be insulated where they cross copper tracks, to prevent shorting.

Because you have to solder on both sides of the PCB you cannot use ICs sockets (unless they are wire-wrap types). The exception here is the HD0165 where all connections are on the copper side. Note also that the HD0165 is not CMOS or MOS and requires no special handling.

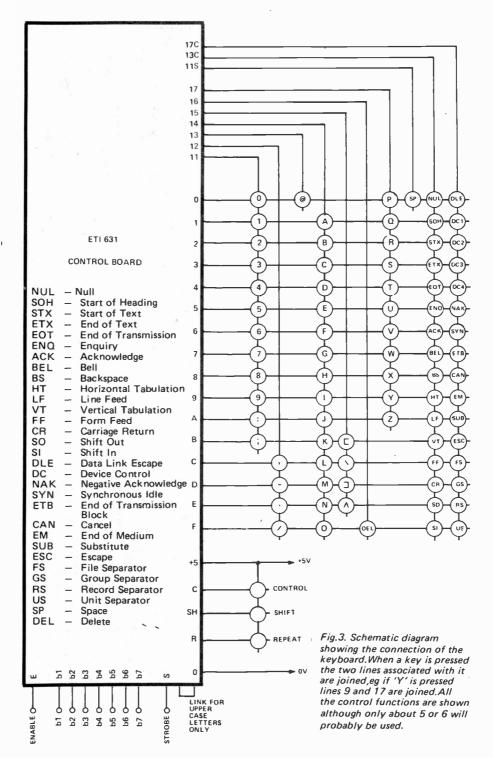
To make wiring easier mark the keys on the underside of the keyboard, to indicate what functions they represent. Now using "solderable" enamelled wire join the points as given in Table 1. The connection from the control board is also given and this should be made

-Parts	List-
-Parts	List.

Resistors all ½ W 5% R1-R7 R8-R12 R13 R14-R16 R17	1 k 10 k 1 k 2k2 10 k
R18	1 M
R19	220 k
R20	82R
R21-R28	10 k
R29	100 k
R30	220 k
R31	100 k
R32	10 k
R33	2k2
Capacitors C1 C2 C3 C4	100 n polyester 4μ7 25 V 1 μ 25 V 1n0 polyester
Semiconductors D1-D3 Q1-Q8 Q9-Q11 Q12-Q14	1N914 BC558 or BC108 BC548 or BC178 BC558 or BC108
Q15	BC548 or BC178
Q16	BC558 or BC108
Q17, 18	BC178
Integrated Circuits	4025 (CMOS)
IC1	4002 (CMOS)
IC2, 3	4001 (CMOS)
IC4	4016 (CMOS)
IC5, 6	NE555
IC7	HD0165

CONTROL CARD	KEYBOARD SEQUENCE	CONTROL KEYBOARD CARD SEQUENCE
0 1 2 3 4 5 6 7 8 9 A B	0 P @ SP 1 Q A 2 R B 3 S C 4 T D 5 E U 6 F V 7 G W 8 H X BS 9 I Y TAB : J Z LF [ ; K ESC	11 : ; 0 9 8 7 6 5 4 3 2 1 12 - / . , 13 @ 14 L O I K J M N H B G F C D E A 15 \(\Lambda\)\(\Lambda
D E F	Λ N DEL Q /	TABLE 1 ETI 631 How to wire up the keyboard

# -SYSTEM 6S



using normal hookup wire. The control functions can be wired between the points given either by taking two wires back to the control board or finding the same wire, if previously used, on the keyboard and linking across.

We have not described a housing for the unit as it will probably be mounted along with the VDU and UART (possibly under a TV set).

However, the control card can

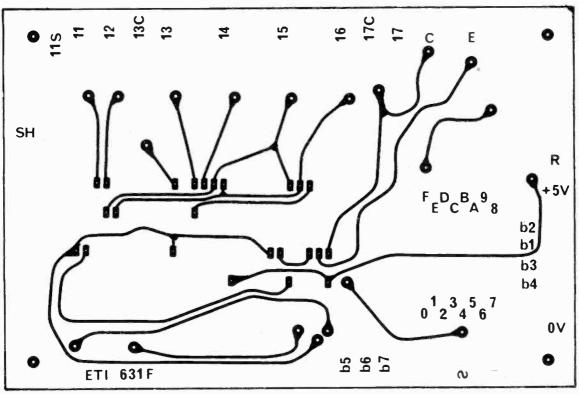
mount under the keyboard by spacing it up slightly. It may be necessary to have a piece of metal (Bacofoil, etc) under the keyboard/control card, connected to OV. (To prevent 50Hz pickup into the wiring to the keyboard.) The effect of this is unwanted outputs from the strobe or non-operation of the strobe output.

To supply the unit 5V at 50mA is needed. To enable the keyboard a

low (OV) is needed on that input. The data output are positive logic (ie, "1" is +5V) and the strobe output is active low.

Connecting the keyboard to a hex display gives an easy check that all wiring is correct. The list in Table 1 gives the character the access, the ASCII code, and the hex code. Alternatively 7 LEDs can be connected (cathode to OV) across the outputs





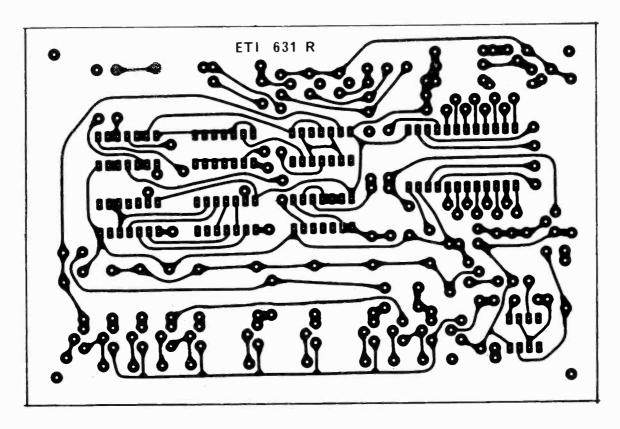


Fig.4. Printed circuit layout(both sides) Full size 150 x 100 mm.



Т	Λ	В	•	E	2
	м	$\mathbf{n}$	L.		

FUNC TION	ACCESS	ASC II CODE	HEX FUNC	ACCESS ASC H CODE	нех
		b5 b6 b3 b3	CODE	67 65 65 64 61	CODE
NUL SOH STX EOT EACK BEL BS LF CR SO SI	CTRL @ CTRL A CTRL B CTRL C CTRL C CTRL F CTRL G CTRL I CTRL I CTRL J CTRL J CTRL L CTRL M CTRL M CTRL M	0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	0 0 0	© 1 0 0 0 0 0 0 0 SHIFT A 1 0 0 0 0 0 0 1 SHIFT B 1 0 0 0 0 1 0 0 SHIFT F 1 0 0 0 0 1 1 0 0 SHIFT B 1 0 0 0 0 1 1 1 0 SHIFT B 1 0 0 0 0 1 1 1 1 SHIFT B 1 0 0 0 1 1 0 0 SHIFT B 1 0 0 0 1 0 0 1 1 1 1 SHIFT B 1 0 0 0 1 0 0 1 SHIFT B 1 0 0 1 0 0 0 1 SHIFT B 1 0 0 1 0 0 0 1 SHIFT B 1 0 0 1 0 1 0 1 SHIFT B 1 0 0 1 1 0 1 0 0 SHIFT B 1 0 0 1 1 1 1 0 1 SHIFT B 1 0 0 1 1 1 1 0 0 SHIFT B 1 0 0 1 1 1 1 0 0 SHIFT B 1 0 0 1 1 1 0 0 SHIFT B 1 0 0 1 1 1 1 0 0 SHIFT B 1 0 0 1 1 1 1 0 0 SHIFT B 1 0 0 1 1 1 1 0 0 SHIFT B 1 0 0 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	4 0 1 4 2 4 3 4 4 5 6 4 7 4 8 9 4 4 B C D E F
DLE DC1 DC2 DC3 DC4 NAK SYN ETB CAM SUB ESC FS GS RS US	CTRL P CTRL Q CTRL R CTRL S CTRL T CTRL U CTRL W CTRL X CTRL X CTRL Z CTRL Z CTRL CTRL \ CTRL \ CTRL \	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	1 0 P Q R 1 1 2 S T U V V X X Y 1 1 8 S T 1 D C C C C C C C C C C C C C C C C C C	SHIFT P 1 0 1 0 0 0 0 0 0 SHIFT Q 1 0 1 0 0 0 0 1 1 SHIFT R 1 0 1 0 0 0 1 1 0 SHIFT R 1 0 1 0 0 0 1 1 0 SHIFT V 1 0 1 0 1 0 1 0 1 SHIFT V 1 0 1 0 1 0 1 1 SHIFT V 1 0 1 0 1 0 1 1 SHIFT X 1 0 0 1 1 0 0 1 SHIFT X 1 0 1 1 1 0 0 0 1 SHIFT X 1 0 1 1 1 0 0 1 SHIFT X 1 0 1 1 1 0 0 1 SHIFT X 1 0 1 1 1 0 0 1 SHIFT X 1 0 1 1 1 0 1 0 1 SHIFT X 1 0 1 1 1 0 1 0 1 SHIFT X 1 0 1 1 1 0 1 0 1 SHIFT X 1 0 1 1 1 0 0 1 SHIFT X 1 0 1 1 1 0 1 0 1 SHIFT X 1 0 1 1 1 0 1 0 1 SHIFT X 1 0 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	0 1 2 3 4 5 6 7 8 9 A B C D E F
SP::#\$%&.():-/	SHIFT 0 SHIFT 1 SHIFT 2 SHIFT 3 SHIFT 4 SHIFT 6 SHIFT 6 SHIFT 7 SHIFT 7 SHIFT 9 SHIFT :	0 1 0 0 0 0 0 0 0 0 1 0 1 0 0 0 1 0 0 1 0 0 0 1 0 1 0 0 0 0 1 1 0 0 0 1 1 0 1 0 1 0 1 0 1 1 0 1 0 1 0 1 0 1 1 1 0 1 0 1 0 1 1 1 1 0 1 0 1 0 1 1 1 1 0 1 0 1 1 1 1 1 1 0 0 1 1 1 1 1 1 1 1 1 0 0 1	2 0 a a b c c d d d e c c d d e c c d d e c c d d e c d d e c d d e c d d d e c d d d e c d d d d	SHIFT @ 1 1 0 0 0 0 0 0 0 1 A 1 1 0 0 0 0 0 1 1 B 1 1 0 0 0 0 1 1 0 0 0 0	0 1 2 3 4 5 6 7 8 9 A B C D E F
0 1 2 3 4 5 6 7 8 9 \ \=\!\!\!\}?	0 1 2 3 4 5 6 7 8 9 : : SHIFT : SHIFT : SHIFT /	0 1 1 0 0 0 0 0 0 0 0 1 1 0 0 0 1 1 0 0 1 1 0 0 0 1 1 0 0 1 1 0 0 1 1 0 0 1 1 0 0 1 1 0 0 1 1 0 0 1 1 1 0 0 1 1 1 0 0 1 1 1 0 0 1 1 1 1 0 0 1 1 1 1 0 1 1 0 1	3 0 P 3 1 r 3 2 s 3 3 4 u 5 3 6 w 7 3 8 y 7 8 7 x 8 9 9 y 7 8 9 9 y 7 8 9 9 y 7 9 9 9 9 9 9 9 9 9 9 9 9 9 9 9 9 9 9 9	P 1 1 1 1 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	7 0 1 2 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7

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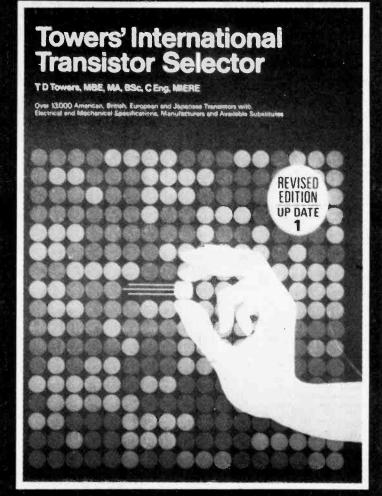
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LINEAR IC PRINCIPLES EXPERIMENTS AND PROJECTS
E. M. Noll
An introduction toyone of electronics most exciting devices



Our Birthday Cake was made for us by DAVID PITHERS, 89 KENWORTHY ROAD, HACKNEY, LONDON E.9.

#### VIDEOMASTER SUPERSCORE

A SIX game unit. It will play football, tennis, squash, solo, A well-proven design this. A very reliable device, and one The kit is very easy to assemble - and even easier to play! fun of building it yourself! (Worth £11.80) (Worth £24.95)

#### P.B. ELECTRONICS 'GIFT BOX'

Superbly presented this contains one u-Dec, one T-Dec, their equivalent 'Blob Boards', pin-leads by the hundred, and four IC holders to plug to the Decs. Everything's included, and everything has its place. The case alone is worth winning, never mind the contents! (Worth £25.00)

#### SPARKRITE IGNITION

and two rifle games (with an optional extra). All with that as a kit is easily and quickly constructed. Offers all the variable ball speed, two angle action, and full sound effects. advantages of electronic ignition systems, along with the

#### **NOVUS LCD WATCH**

Exactly as featured in our phenomenally successful reader offer. A five function, continuous display electronic watch. Along with the accuracy of the quartz crystal timebase, and the superb contrast of the display, goes a seconds readout, and day-date (U.S. style) format. (Worth £16.95)

#### **VEROWIRE KIT**

This is a fairly new venture for the 'board people' - a complete wiring-pen outfit, including pen and wire-reels, magnifier, wiring harnesses, pins, wire-cutters, pin insertion tool, and dil prototyping board. A definite essential if you do a fair bit of digital work, and very useful indeed for prototyping ANY kind of circuit. (Worth £18.15)

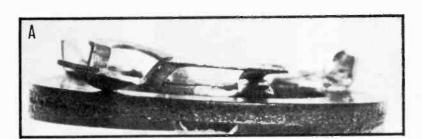
- RULES OF THE GAME 1. Any persons in any way connected with the competition are ineligible to enter. This includes employees of firms manufacturing the prizes, and of Modmags Ltd.
- The decision of the judges shall be regarded as final.
- 3. The prizes will be awarded to the first five correct entries drawn after the closing date, which shall be April 30th. No correspondence can be entered into regarding the competition.
- 4. All entries must be on the coupon from the magazine. Photostats are not acceptable.
- 5. Results will be published in the July issue of ETI, and winners notified by post.

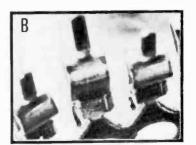
# THDAY COMPETITION

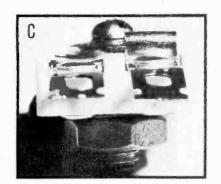
A bit of a change this time. Below are eight photographs of electronic components or equipment. All you have to do is to tell us what they are. Of course we've taken them from as obscure an angle as possible — just for fun. None of the items depicted is as rare as it looks; you will certainly have seen them all many times from many angles. Fill in the answers on the coupon. For instance, if you think A is an atomic tea-strainer and butterfly mincer, write that next to A on the coupon. One hint: A is not an atomic tea-strainer.

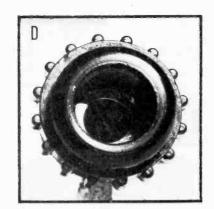
Competition closes 30/4/77 and the results will be announced in the July issue of ETI. Good luck

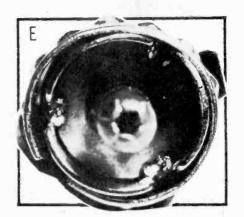
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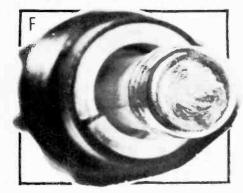


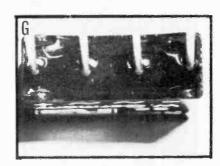


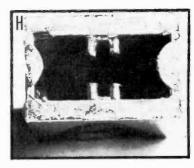










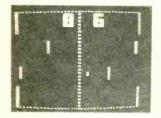


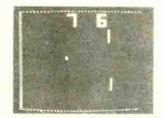
To: BIRTHDAY COMPETITION, ETI-	Magazine 25-27 Oxford Street, London,	W1R 1RF. Closing date: 30/4/1977.
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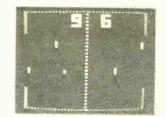
# Clectronics today international

What to look for in the May issue-on sale April 1st.

# At last: a single-chip TV game!









We've done it again! Next month we're presenting a really comprehensive DO-IT-YOURSELF T.V. game design. Our unit plays football, tennis, squash and a practice game. All have two speed action, variable reflection angle, and naturally on screen scoring and full sound effects. Boundaries are clearly shown, and the size of bat can be altered. If you think you can play T.V. tennis try

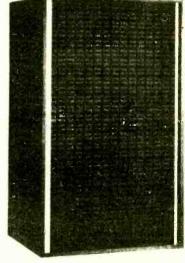
playing our game with small bats and variable angle!

The game is based on the G.I. AY-3-8500 chip, the retail price of which is £15. However, for those of you who will be building this project, Maplin Electronics (see back cover for address) are making a special offer of £8.99 inc. until April 18th. (Trade enquiries are welcomed!) A full kit of parts will be available upon publication of the article.

# Hi-Fi Speaker

# Enclosure

Our first venture into speaker design is a large bookshelf/small, free-standing, high quality enclosure. We avoided the usual 'megalithic monster' syndrome, and paid very special attention to the crossover and enclosure design instead — both sadly neglected areas. The result is an enclosure which has a performance that totally belies its size, and is very simple (and reasonably cheap!) to construct. The crossovers are available ready built off the shelf — so no coil winding — and full woodwork details are given in the article.



# **SHORT CIRCUITS**

#### SIGNAL INJECTOR/TRACER

A handheld unit with its own amplifier and oscillator. The amp is usable separately from the oscillator, and provides very reasonable quality for circuit checking.

#### **METRONOME**

A two-transistor design with a wide range from several seconds to several per second — and we include a fixed position with a beat rate that shows some evidence that stammerers can be helped.

# $I^2L$

We've heard about I'L, you've heard about it, but until now it has been almost impossible to find out what this new technique is, where it can be used and what advantages it has over TTL and/or CMOS.

In the May issue we tell all.

# The Valves that Won the War

The last world war was the first fought using technology and the Allies eventually outstripped the Axis powers and that led to our victory as much as the cheerful Tommy and the resiliant Ivan. Radar was perhaps the one most important development but this only became practical and portable with the development of the klystron.

Features mentioned on this page are in an advanced state of preparation but circumstances, including late developments may affect the final contents.

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	.35	2N3704		40362	0.48	BC161	0.50	BD135	0.37		.25						
	.30	2N3705	0.15	40363	1.20	BC167	0.12	BD136	0.37		.38	CA302D		LM3301N		TAA661B	
	.62	2N3706	0.16	40406	0.58	BC168	0.12	BD137 BD138	0.38		.36	CA3020A		LM3302N		TAA700	
	.55	2N3707	0.18	40407	0.45	BC169	0.12	BD138	0.40			CA3028A		LM3401		TAA930A	
	.24	2N3708	0.16	40408	0.65	BC170	0.16	BD 140	0.40		.34	CA3028B		LM3900		TAA930B	
	.12	2N3709	0.18	40409	0.65	BC171	0.14	BD239	0.40		.50	CA3030		LM3905	1.60	TAD 100	1.9
	.21	2N3710	0.16	40410	0.65	BC172	0.12	BD 240	0.45		.30	CA3030A		LM3909	0.68	TBA 120	0.6
	.50	2N3711	0.18	40411	2.85	8C177 BC178	0.20	BD241	0.45		.32	CA3045	1.40	MC1035	1.75	TBA400	1.50
	.27	2N3712	1.20	40594	0.75	BC178	0.20	BD242	0.47		.50	CA3046	0.89	MC1303	1.47	TBA500	2.2
	.50	2N3713	2.30	40595	0.85	BC179	0.11	B0243	0.60		.20	CA3048	2.23	MC1304 MC1305	1.85	TBA500Q	
	.80	2N3714	2.45	40673	0.73	BC182L	0.14	BD244	0.62		.20	CA3049	1.66	MC1305	1.85	TBA510	
	.35	2N3715	2.55	AC126	0.37	BC183	0.11	BD245	0.65		.15	CA3052			1.91		
	.30	2N3716	2.80 1.85	AC127	0.44	BC183L	0.14	BD246	0.66		.20	CA3053 CA3080	0.60	MC1312	1.98	TBA520 TBA5200	
	.38	2N3771 2N3772	2.00	AC128	0.37	BC184	0.12	B0529	0.42		.10	CA3080A		MC1327	1.54		1.9
	.25	2N3773	2.90	AC151V	0.35	BC184L	0.14	BD530	0.47		.10	CA3086	0.51	MC1330	0.92		
	.26	2N3773	2.90	AC152V AC153	0.50	BC207	0.12	BDY20	1.13		35	CA3088	1.59	MC1350	0.75	TBA530Q TBA540	
	.60	2N3790	3.10	AC153K	0.49	BC208	0.11	BF115	0.38		.55	CA3089	2.52	MC1351	1.20	TBA5400	
	.60	2N3791	3.10	AC176	0.40	BC212	0.14	BF117	0.70		.35 .	CA3099	3.80	MC1352	0.97	TBA550	
	.35	2N3792	3.50	AC176K	0.60	BC2121	0.17	BF121	0.55	MJ491 1.	.85	CA3130	0.94	MC1357	1.45	TBA5500	
	.38	2N3794	0.20	AC187K	0.55	BC213	0.14	BF123	0.55	MJ2955 1.	.25	LM301A	0.65	MC1458	0.91	TBA5600	
	.60	2N3819	0.36	AC188K	0.55	BC21BL	0.16	BF152	0.25	MJE340 0.	.58	LM301N		NE555	0.53	TBA570	
	.33	2N3820	0.38	AD161	0.85	BC214	0.16	BF153	0.25	MJE370 0.	.58	LM304	2.45	NE556	1.05	TBA5700	
2N2218A 0		2N3823	0.75	AD162	0.85	BC214L	0.17	BF154	0.25		.60	LM307N		NE565	1.20	TBA641B	
	.30	2N3904	0.21	AF106	0.55	BC237	0.14	BF159	0.35	MJE520 0.	.45	LM308C	1.82	NE566	1.65	TBA651	1.8
2N2219A 0		2N3906	0.22	AF109	0.75	BC238	0.12	BF160	0.30		.65	LM308N	1.17	NE567	1.80	TBA 700	1.5
	.35	2N4036	0.67	AF124	0.65	BC239	0.15	BF161	0.60	MJE2955 1.		LM309K	2.10	SASS60	2.50	TBA 7000	
	.22	2N4037	0.55	AF125 *	0.65	BC251	0.15	BF166	0.40	MJE3055 0.		LM317K	3.00	SAS570	2.50	TBA 7200	
2N2221A 0		2N4058	0.20	AF126	0.65	8C253	0.22	BF167	0.38		.35	LM318N	2.25	76001N	1.57		1,9
	.25	2N4059	0.20	AF127	0.65	BC257A	0.17	BF173	0.38		.40	LM323K	6.40	76003N	2.55	TBA750Q	
2N2222A 0.		2N4060	0.20	AF139	0.69	BC258A	0.17	BF177	0.30		.45	LM339N	1.75	76008K	2.50	TBA800	1.2
2N2368 0.	.25	2N4061	0.17	AF186	0.50	BC2598	0.18	BF178	0.35		.30	LM348N	1.91	76013N	1.70	TBA810	1.10
2N2369 0		2N4062	0.18	AF200	0.70	BC261A	0.21	BF179	0.35		.23	LM360N	2.75	76013ND	1.57	TBA820	1.0
2N2369A 0	.25	2N4126	0.17	AF239	0.74	BC262B	0.19	BF 180	0.40		.24	LM370N	3.00	76018K	2.50	TBA920	1.7
	.75	2N4289	0.20	AF240	0.98	BC263C	0.24	BF181	0.40		.35	LM371N	2.25	76023ND		TBA9200	
2N2647 1.	.40	2N4919	0.65	AF279	0.80	BC300	0.45	BF182	0.45		.24	LM372N	2.15	76033N	2.55		1.6
	.36	2N4920	0.70	AF280	0.85	BC301	0.45	BF183	0.45		.24	LM373N	2.25	76110N	1.46	TCA160C	
2N2904A 0.		2N4921	0.50	BC107	0.15	BC303	0.60	BF 184	0.38		.50	LM374N	2.25	76114N	1.87	TCA160B	
	.37	2N4922	0.55	BC108	0.15	BC307	0.20	BF185 8F194	0.35		.56	LM377N	1.75	76116N	2.06	TCA270	
2N2905A 0.		2N4923	0.70	BC109	0.15	BC308	0.18	BF 195	0.13		.55	LM378N	2.25	76131N	1.30	TCA 280A	
2N2906 0		2N5190	0.60	BC113	0.17	BC309C	0.25	BF 196	0.14		.60	LM379N	3.95	76226N	1,94	TCA290A	
2N2906A 0		2N5191	0.70	BC115	0.19	BC317 BC318	0.14	BF 197	0.17		.45	LM380-8		76227N	1.51	TCA420A	
2N2907 0.		2N5192	0.75	BC116	0.19			BF198	0.18		.50	LM380N	0.98	76228N	1.75	TCA730	3.2
2N2907A 0		2N5195	0.90	BC116A	0.20	BC327 BC328	0.20	BF200	0.35		.50	LM381A	2.45	76530N	0.91	TCA 740	2.7
	.15	2N5245	0.35	8C117	0.22	BC328	0.19	BF225J	0.25		.80	LM381N	1.60	76532N	1.50	TCA750	2,3
	.13	2N5294	0.40	8C118	0.16	BC338	0.19	BF244	0.35		.90	LM382N	1.25	76533N	1.30	TCA760	1.3
	.55	2N5295 2N5296	0.40	BC119 BC121	0.30	BC547	0.12	BF245	0.34		.50	LM384N	0.80	76544N 76545N	2.09	TCA800 UAA170	3,1
	.30	2N5298	0.40		0.45	BC548	0.12	BF246	0.75		.35	LM386N			1.44		2.0
	.70	2N5298 2N5447	0.40	BC132 BC134	0.30	BC549	0.13	BF 254	0.24		.70	LM387N LN388N	1.05	76546N 76550N	0.41	UAA180	2.0
		2N5448	0.15	BC135	0.15	BCY30	1.03	BF255	0.24		.80	LM389N	1.00	76552N	0.65	DIL	
	.25	2N5449	0.19	BC136	0.19	BCY31	1.06	BF257	0.37		.60	LM702C	0.75	76570N	2.08	SOCK	
2N3391A 0		2N5457	0.32	BC 137	0.14	BCY32	1,70	BF258	0.45	TIP30c 0.	.65	LM709C	0.65	76620N	1.10	8 pin	0.1
	.16	2N5458	0.33	BC140	0.40	BCY33	1.00	BF259	0.49		.66	LM7090	0.45	76650N	1.10	14 pin	0.1
	.15	2N5459	0.33	BC141	0.45	BCY34	1.20	BF459	0.45		.75	LM710C	0.60	76660N	0.60	16 pin	0.1
	.15	2N5484	0.34	BC142	0.30	BCY38	2.00	BFR39	0.28		.10	LM710N	0.60	76666N	0.92	22 pin	0.3
	.88	2N5486	0.38	BC143	0.30	BCY42	0.60	BFS21A	2.60		20	LM723C	0.85	TAA301A		24 pin	0.3
	.64	2N6027	0.53	BC147	0.12	BCY58	0.25	BFS28	1.04		.85	LM723N	0.75	TAA320A		28 pm	0.4
	.85	2N6101	0.65	BC148	0.12	BCY59	0.25	BFS61	0.30		.95	M741C	0.65	TAA350A	2.48	40 pm	0.5
	.35	2N6107	0.42	BC149	0.13	BCY70	0.25	BFS98	0.27		.65	LM 741N	0.50	TAA521.	1.00	,	
	.16	2N6109	0.42	BC153	0.27	BCY71	0.26	BFX29	0.38		.55	LM741-8		TAA522	1.90		
2N3638A 0		2N6121	0.38	BC154	0.27	BCY72	0.24	BFX30	0.40	TIS43 0.	.30	LM747N		TAA550	0.60	EXPR	ESS
	.30	2N6122	0.41	BC157	0.12	BD115	1.20	BFX84	0.40			LM748-8	0.50	TAA560	1.60	M.O SE	
2N3641 0	.20	2N6123	0.43	BC158	0.11	BD116	1.20	BFX85	0.41	MANY OTHI		LM748N	0.50	TAA570	2.30	BY RET	
	.17	2N6126	0.45	BC159	0.14	BD131	0.51	BFX87	0.40	TYPES IN		LM1800	1.76	TAA611B			
2N3703 0.	.15	40361	0.45	BC160	0.50	BD132	0.54	BFX88	0.40	STOCK			1.92	TAA621		OF PC	151
	- 4					2						LM1828	1.75	TAA661A	1.32	1	

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CD4014	1.15	CD4030	0.64	CD4518	1.43
CD4015	1.15	CD4031	2.53	CD4520 '	1.43
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SN7405	0.26		0.39	SN 7447	1.17	SN7475	0.58	SN7492	0.61	SN74123	0.58	SN74163	1.41	SN74192	1.62
SN 7406	0.74	SN7425	0.39	SN7448		SN7476	0.51	SN7493	0.61	SN74141	1.03	SN74164	1.23	SN74193	1.62
SN7407	0.74		0.39	SN7450		SN 7480	0.45		0.74	SN74145	1.06	SN74165	1.23	SN74196	1.17
SN7408	0.29	SN7430	0.21	SN7451		SN7481		SN7495	0.78		1.66	SN74167	3.70	SN74197	1.17
SN7409	0.29	SN7432	0.39	SN7453		SN7482	0.67	SN7496	1.03	SN74151	1.11	SN74174	1.52	SN74198	2.93
SN 7410	0.21	SN7437	0.55	SN7454	0.21	SN7483	1.33	SN74100	1.15	SN74153	1.11	SN74175	1.35	SN74199	2.93
Name of the						literature in				10 miles 17 miles					

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# ETI DATA SHEET

#### MK 50361-50362 ALARM CLOCK

**MOSTEK** 

THESE ARE MOS CIRCUITS, using depletion-load ion implantation process. Both chips work in 12 or 24 hr mode and will drive displays directly. Features included are: 10mA/segment LED drive; low voltage backup (9v); forward or reverse time setting; intensity control; non-MPX display; 24hr alarm in three modes — tone, radio or tone plus radio; variable 'snooze' (1-59 mins); count hold; summer/winter time switch; clear on switch on; leading zero suppression in 12hr mode. ON 50362 ONLY: Four Year Calendar; choice of date display format; second alarm time.

The snooze inhibits an activated alarm for 10min periods. The 'sleep' will activate the radio for between 1-59 mins (adjustable).

#### **POWER FAILURE**

At the occurrence of power failure, the digits will flash at 1Hz to indicate incorrect time displayed. Set 2 (forward) switch should be closed (once power is restored). This mode is triggered when the colon output goes to  $V_{\rm SS}$  and may occur at initial power-on.

#### **DISPLAY MODE**

The input is three state. Operation is as follows:

#### Display Input Mode

V<sub>SS</sub> Alarm time displayed Open Time displayed V<sub>DD</sub> Seconds displayed

When in alarm mode, the time displayed is that to which the alarm is set. It may be altered by use of the time set procedure, given below.

#### **SETTING**

The setting mode allows either a forward setting *or reverse setting* of the display. The setting inputs are:

Set 1 or	
Set 2 Input	Mode
$V_{SS}$	Forward Set
Open	
$V_{DD}$	Reverse Set

When either the set 1 input or set 2 input is connected to  $V_{SS}$  the display will increment. Connecting the input to  $V_{DD}$  will decrement the display. When the display is not being set, the inputs should be left open. The set 1 input changes the hours digits at two counts per second. The set 2 input changes the minute digits at two counts per second. Carrys or borrows are not allowed during time setting.

#### **DISPLAY FORMAT**

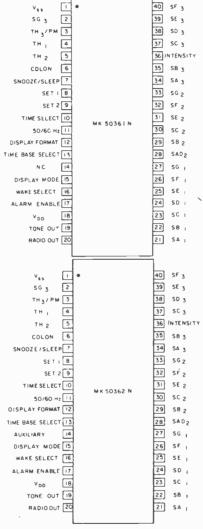
The display format is used to select a 12 hour display, 24 hour display or to blank the display. The connections are

# Display Format Input VSS 12 Hour Open Blank Display VDD 24 Hour

In the 12 hour mode, the hours digits will display time in 12 hour format with a PM output. When the input is connected to  $V_{\rm DD}$ , the format will be 24 hour time. In the blank mode the segment outputs will float, allowing wire-or conditions.

#### COLON

In normal operation, the colon flashes at one Hertz rate for an activity indicator. The colon output conducts to  $V_{SS}$  with a 50% duty cycle in the 60Hz mode and 40% duty cycle in the 50Hz mode. The colon is off when displaying calendar, seconds or sleep time.



#### **INTENSITY**

The intensity input regulates the current of the segment outputs and colon output.

The intensity input regulates the current of the segment outputs and colon output. Over a range from  $3K^+$  to  $30K^+$  for the intensity resistor ( $R_{\rm SENSE}$ ), the following equation may be used to predict segment current ( $I_{\rm SEG}$ );

Segment current is relatively independent of the voltage on the segment pins, therefore the LED display voltage supply need not be well filtered. For minimum value of R<sub>SENSE</sub> (intensity control resistor), care should be taken to insure total circuit power dissipation does not exceed safe operating limits

#### **AM/PM OUTPUT**

When in the 12 hour operating mode, the TH3/PM output conducts to  $V_{SS}$  when active. The indicator will change states when the hours change from 11 to 12. When in the 24 hour mode this output drives segments A, D & G on the most significant digit. Consideration must be given to wiring the display to accommodate both 12 hour and 24 hour operation using the same display.

#### TIME SELECT (Daylight saving time)

Connection of the time select pin to  $V_{SS}$  will advance the time by one hour. By opening the connection, the time will return to the original time. Connecting this input to  $V_{DD}$  will reset the clock to the Power Up mode.

#### ALARM ENABLE/WAKE SELECT

The alarm can operate in three modes according to the voltage on the wake select pin. The states are defined as

Wake Input	Mode
$V_{SS}$	Tone
Open	Radio
$V_{DD}$	Radio followed
	by Tone

The Alarm enable pin enables alarm 1 when connected to  $V_{SS}$ . If it is left open it will disable the alarm due to an internal resistor. When the alarm occurs it may be disabled and immediately re-enabled and will activate 24 hours later at the alarm time. The alarm will self-disable after one hour of operation. The output tone will be in the range of 200 to 1000Hz and conducts to  $V_{SS}$  8.3% of the time at a 1Hz rate.

Radio out, when activated by either the alarm or sleep function, will conduct to  $V_{SS}$ .

When the radio followed by tone mode is selected, radio out conducts to V<sub>SS</sub> at alarm time. Eight minutes later the tone output will be enabled. Both remain on until inhibited by the alarm enable control, snooze, or the automatic alarm reset which occurs after one hour.

#### SNOOZE/SLEEP INPUT

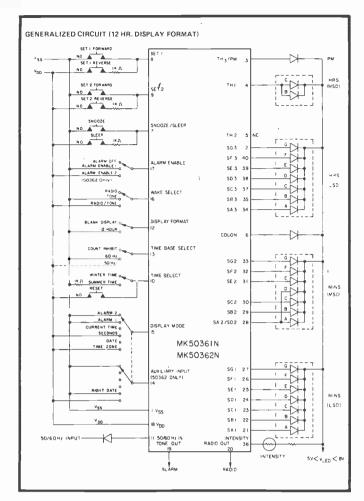
The snooze and sleep inputs use a single pin to select snooze or sleep. The connections are:

Snooze/Sleep	Mode
$V_{SS}$	Snooze
Open	No Change
$V_{DD}$	Sleep

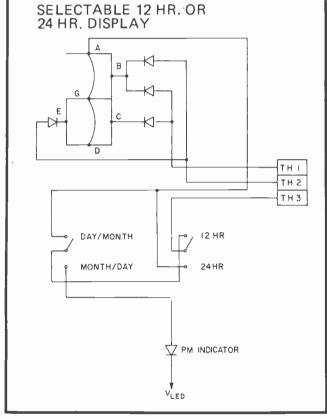
The Snooze feature will temporarily turn off an activated radio and tone outputs to allow an additional 10 minutes' sleep. Momentarily connecting snooze to  $V_{SS}$  will activate the snooze. If left open an internal pull-down resistor will maintain the snooze feature inoperative.

Connection of the pin to  $V_{DD}$  will display the sleep time in minutes in the minutes digits. The time will start at 10 minutes, and the set 2 input is used to set the sleep time. Radio out will conduct to  $V_{SS}$  for the amount of time set. After the time decrements to zero radio out will turn off, provided that the Snooze/Sleep input is not being held at  $V_{DD}$ . If the snooze is active, the sleep input will

If the snooze is active, the sleep input will reset the snooze function. The snooze input will reset an active sleep time.



PARAMETER	Min	Max	Units
Operating Voltage, VDD	-12	-16	Volts
Standby Voltage, VDD	-8.0	-12	Volts
Input Logic Levels: Set 1, Set 2, Display Mode, Snooze/Sleep, 50/60Hz, Display Format, Alarm Enable, Wake Select, Time Select, Time Base Select, Auxiliary	V <sub>SS</sub> -1.0	Vss	Volts
	VDD	V <sub>DD</sub> + 1.0	Volts
Intensity Control	3	30	K-OHMS
Segments, Colon		-26	Volts



THESE CHIPS ARE AVAILABLE FROM PRONTO ELECTRONIC SYSTEMS LTD. 645-647, HIGH ROAD, SEVEN KINGS, ESSEX, IG3 8RA. PRICE-£8.00 inc.

DISPLAY FONT

#### **FUNCTION SETTING**

The displayed function is set using the set inputs while the appropriate function is displayed using the display mode input and auxiliary input. The set 1 input changes the hours or month digits at two counts per second. The set 2 input changes the minute or date digits at two counts per second. Carrys or borrows are not allowed during setting except for an illegal month date combination.

The MK 50362N contains an auxiliary input which allows the selection of additional features. All other functions operate like the MK 50361N.

The additional features are selected by using the Display Mode Input and the Auxiliary Input. The selection is:

	Display Mode	Auxiliary
Function	Input	Input
Alarm 1 Set	$V_{SS}$	Open
Current Time	Open	Open
Seconds	VDD	Open
Month Date	Open	$V_{SS}$
Date Month	Open	V <sub>DD</sub>
Alarm 2 Set	$V_{SS}$	$V_{SS}$
Time Zone	VDD	V <sub>DD</sub>

#### **MONTH-DATE CALENDAR**

The calendar is a four year calendar. Connecting the Auxiliary input to  $V_{SS}$  will display a Month-Date format. A Date-Month format can be selected by connecting the Auxiliary input to  $V_{DD}$ . The display mode input must remain open.

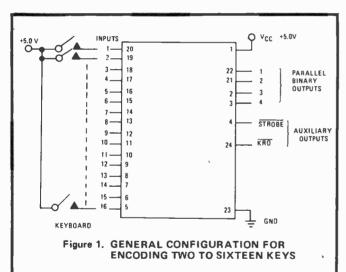
#### SECOND ALARM TIME

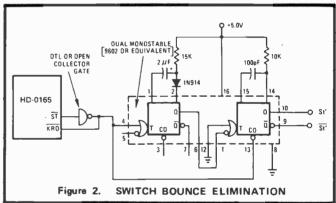
The second alarm time can be displayed by connecting the Display Mode input and the auxiliary input to  $V_{SS}$ . To enable the alarm, the Alarm Enable pin should be connected to  $V_{DD}$ . Alarm 2 will not have an automatic shutoff. Disabling either alarm will reset snooze.

The HD-0615 Keyboard Encoder is a 16 line to four-bit parallel encoder intended for use with manual data entry devices such as calculator or typewriter keyboards. In addition to the encoding function, there is a Strobe output and a Key Rollover output which energises whenever two or more inputs are energised simultaneously. Any four-bit code can be implemented by proper wiring of the input lines. Inputs are normally wired through the key switches to the +5.0V power supply. Full typewriter encoding up to eight bits can be accomplished with two Encoder circuits by the use of double pole key switches or single pole switches with two isolation diodes per key. Outputs will interface with all popular DTL and TTL logic families. The circuit is packaged in a hermetic 24-pin dual in-line package and operates over the temperature range of 0°C to + 75°C.

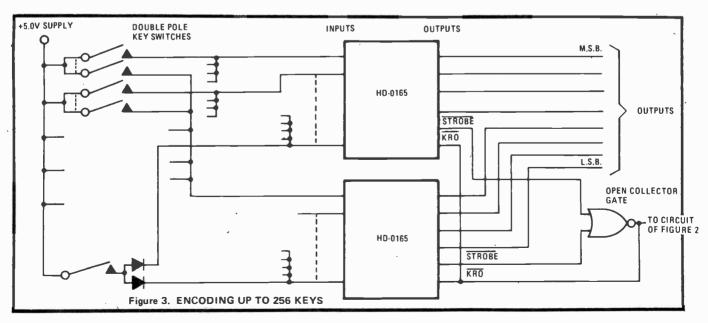
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				LIMITS			UNITS	
	PARAMETER		SYM.	MIN.	TYP.			TEST CONDITIONS
	Input Current	1	I <sub>H</sub>			17	mA.	V <sub>IN</sub> = +5 0V
J.C.	Output Voltage	0	v <sub>ol</sub>		+0.2	+0.4	v	V <sub>IH</sub> = +4.5V 1 <sub>O L</sub> = 10mA V <sub>IH</sub> = +3.5V 1 <sub>O L</sub> = 2.2mA
		1	VOH	+2.4	+40			VIL = Open Circuit, IOH = -240 L.A.
	Power Supply Current	Operating	1 <sub>CC</sub>			52	mA	One input at +5.25V
		Maximum	<sup>I</sup> CCM			88	mA	All Inputs at +5.25V
A.C.	Skew Time (Note 1)		<sup>T</sup> SK		80	200	ns	T <sub>Case</sub> = 25°C V <sub>CC</sub> * V <sub>IN</sub> = +5.0V C <sub>L</sub> < 50pF





THE ENCODER IS AVAILABLE FROM PRONTO ELECTRONICS, ADDRESS PAGE 39, FOR £8.10 inc.



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pads, total area 9½ x 7½in.)

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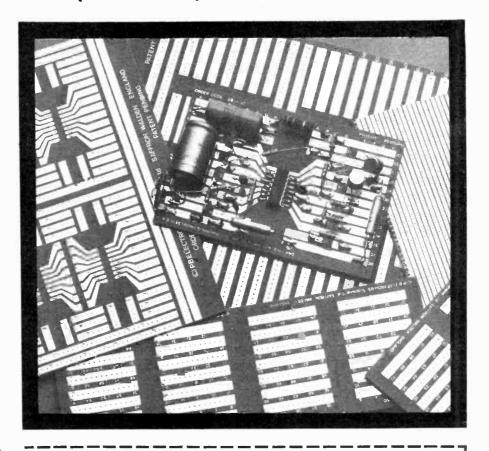
Comprises one of each:

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ZB5D [total area 3% x 2%in.]

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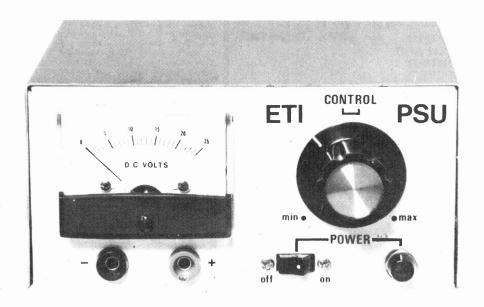
The meter should be looked upon as an optional 'luxury' extra. The design will function perfectly without it, and will be nearly a fiver cheaper to construct. Use a descent pot, and calibrate it.

#### CONSTRUCTION

Physically the most difficult part of construction will undoubtedly prove to be drilling the case. Mount all the components to the PCB as shown in the overlay, noting that BR1 and C1 are mounted onto T1, not the board.

The short circuit protection resistor R2 is specified as OR5 (½ ohm) at 5W. You could use two 1R in parallel (at 2.5W each) if you have trouble obtaining the component.

Q2, the output series pass transistor, must be isolated from the case which functions as its heatsink. Also ensure the connections to C1 do not short to the case while you're at it! The heatsinks for IC1 can be bent to any convenient shape or size, the minimum size for which may be taken to be that which we used! There is no DC (output) fuse, as the IC can look after itself better than any fuse!

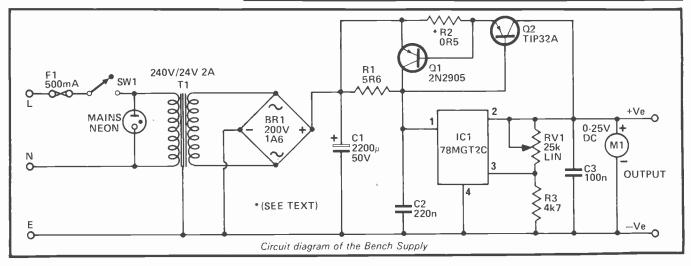


## How it works

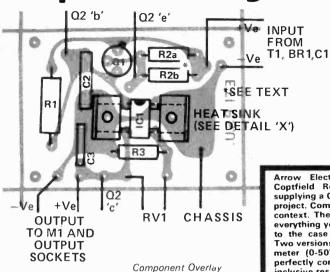
What can we say? IC1 does all the work. C1 is provided to smooth out the full wave rectified DC from BR1. The voltage at the output pin 2 of IC1 is varied by varying the control voltage applied to pin 3.

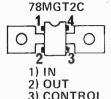
Q2 is used to enable more current to be drawn than IC1 could provide. Unaided, it could supply 500mA. In this circuit loads can draw in excess of 1.5A without the supply shutting down. Should the dissipation of the chip exceed safe limits, the output will be limited.

Short circuit protection for the series pass transistor is provided by R2 and Q1. C2 and C3 provide input and output bypassing, and their retention is recommended to ensure stability in all possible conditions.



Short Circuits

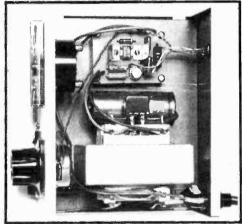




3) CONTROL 4) COMM

ALSO HEAT SINK TABS ARE COMM.

Arrow Electronics Ltd., Leader House, Coptfield Road, Brentwood, Essex, are supplying a COMPLETE kit of parts for this project. Complete here is used in its proper context. The P.C.B. is screen-printed, and everything you need is in there, right down to the case and P.C.B. stand-off pillars! Two versions are available, with or without meter (0-50V not as our prototype, but perfectly compatible) at £12.60 or £16.60 inclusive respectively



Parts List-

RESISTORS

R2

5R6 1W R1 ORS SW (SEE TEXT)

4k7 ½W R3

CAPACITORS

2.200u 50V electrolytic C1

C2 220n polyester СЗ 100n polyester

**SEMICONDUCTORS** 

2N2905 or similar TIP 32A or similar 02 IC1 uA78MG T2C

(positive voltage regulator)

200V 1.6A Bridge Rectifier RR1

**POTENTIOMETER** 

25k Lin. rotary RV1

**SWITCH** 

SW1 Off/On rocker type 3A 240V

**METER** 

0-25V DC panel type

CASE

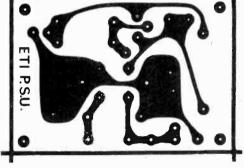
Samos S6 Doram: 984 - 481

TRANSFORMER

240V - 24V 2A type

MISCELLANEOUS

Knob, insulating kit for Q2, fuse holder, fuse, 4mm. red and black sockets, grommet, P.C.B. pillars, nuts, bolts, etc, flex, 3-core mains flex, mains neon.



PCB Foil Pattern (full size)

# FUZZ BOX

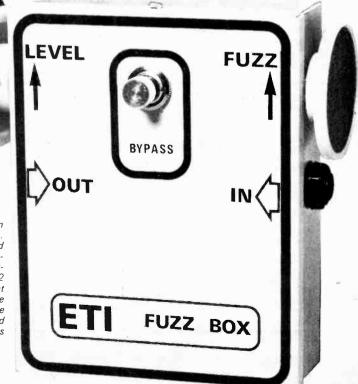
STRANGE AS IT SOUNDS, by far the most popular and sought after effect these days is distortion. More fuzz boxes have been constructed and purchased than any other of the myriad types of guitar 'modifiers' on the market.

Although the aim is basically very simple, some very high prices are being asked, which would tend to make one think (or hope!) that there is more to the principle than meets the soldering-iron. We've kept our's simple: the total cost including board, case and footswitch (£1.50 alone) should be approximately £6.00.

#### CONSTRUCTION

The stereo jack SK2 is used to switch the unit on and off; when a mono jack is inserted it shorts two contacts, and completes the supply circuit. The PP6 battery specified

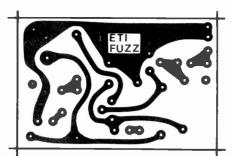
Care has been taken in the mechanical design. The bypass switch and both level and fuzz controls can be foot-operated whilst playing. RV2 can be set so level at fuzz and bypass are equal, however it can be used with RV1 to extend the range of effects possible

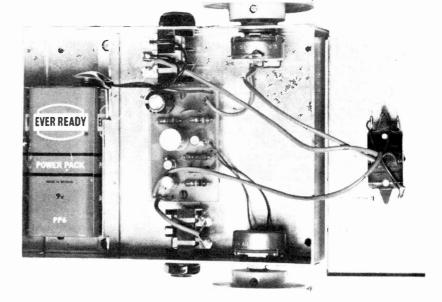


## -How it works-

Transistors Q1 and Q2 amplify the incoming signal, and the gain is such that the input will 'overload' when used with an electric guitar. RV1 adjusts the amount of feedback present, and hence voltage gain.

The output is therefore a 'squared' version of the input signal, the amount of 'squaring' being variable by RV1.





## —Parts List

RESISTORS All ½W 5% R1 39k R2 100k R3 680R R4 5k6 R5 56R

#### CAPACITORS

C1 10u 16V electrolytic C2 100u 16V electrolytic C3 47n ceramic

C4 47n ceramic C4 47u 16V electrolytic

#### SEMICONDUCTORS

Q1,2 BC108 or similar

#### POTENTIOMETERS

RV1 1k Lin. rotary RV2 100k Log. rotary

#### SWITCH

SW1a,b Double pole, change

over, push on/push off, footswitch. Bulgin SM270 type or similar is available from two advertisers: ANCO, 50 Rainsford Rd., Chelmsford - or MAPLIN, P.O.Box 3, Rayleigh, Essex Essex. Approx.price: £1.50

#### SOCKETS

SK1 Mono ¼" jack socket SK2 Stereo ¼" jack socket

#### CASE

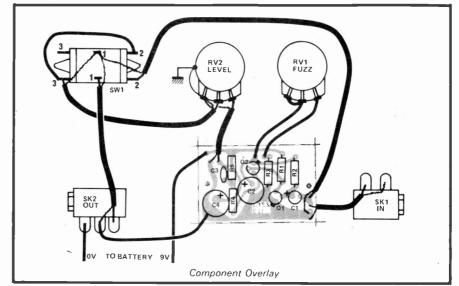
 $6^{\prime\prime}$  x  $4^{\prime\prime}$  x  $2^{\prime\prime}$  approx. Metal type from H.L.Smith (130 x 103 x 50mm.)

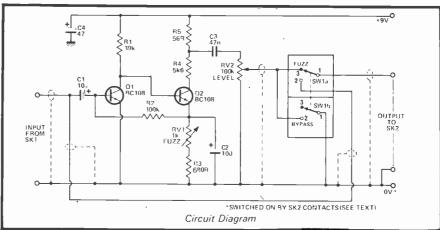
#### **MISCELLANEOUS**

2 large knobs, P.P.6 battery, battery clip, screened wire, P.C. board pillars, nuts, bolts, etc., P.C. board as shown.

APPROX' COST:

£6.00 inc. board and battery.





should last a long, long time, as current drain is quite small. Use screened wire where shown on the circuit; to link SW1 and the output, and the 'bypass' signal with the output.

This switch, SW1, not only switches the signal through the circuit, but also takes the output from RV2 to ground

when in the 'non-fuzz' position, preventing breakthrough of the distorted signal onto the line.

Assemble the PCB in the usual manner; watch the orientation of the transistors lest you get a 'fuzzed' junction. Take the case of RV2 to the 'earthy' side of the pot.

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POP11/20 Processor (Negative bus) with DW08 Bus Converter DF32 32K Disk Drive and Control (Slave drives type DS32 also available if required) 6801 Communications Data Lines PDP9 24K Processor complete with KEO9 Extended Arithmetic, TCO2 DECtape controller + 2 TU55 drives, PCO9 Paper Tape Reader and Punch, AFO18 A/O Converter (8-channel) DA098 I/O Bus LT198 Multi-teletype station.
RPO2 30 Meg, Free-standing moving-head disk drive and control. TU10 Magnetid Tape Units, 9-track 800 bp. rack-mounted TU20 Magnetid Tape Unit, 9-track, 45 ips. TU55 DECtape drives. DF32 Disk Drive and Control. DS32 Slave Drives also available.

DF32 Disch Prive and Control.

DF32 Disch Prive and Control.

CENTRONICS 101 Line Printer and Control.

RK11/RK05 Disk Drive and Control.

RK11/RK05 Disk Drive and Control.

RT02 Alphanumeric Data Entry Olsplay and Control.

DEC Interface and Feature Boards. DC11-AC, DC11-DA, DL11C.

DI11E, KDBE, KLBE, KE11F, KW11, KCBE, MCBEJ, MM11F, MRBE, M7239 G235 KE11 DB11 M8342 G1110/G231.

MRBE M7239 G235 KE11 DB11 M8342 G1110/G231.

H754

DEC Power Supplies H724, 728A, etc.

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control logic Price £295.00.
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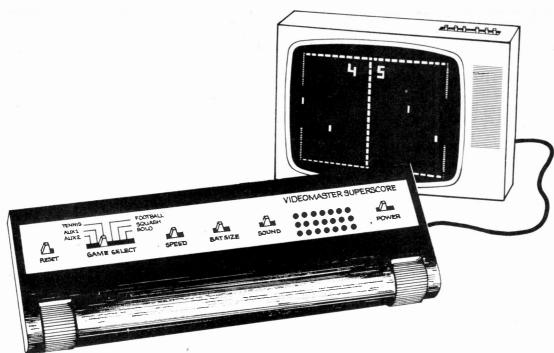
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notepaper.

# WIREWOURD BESISTORS

# PART 9

This month concerns itself with the usually higher powered wirewound resistor variety

These resistors are made by winding a length of resistance wire on a bobbin (usually of ceramic or fibreglass), the ends being anchored to terminations on the ends of the bobbin. Bobbins are usually cylindrical-shaped or flat. The bobbin and element are generally encapsulated in an impervious coat of vitreous enamel — some styles have the whole bobbin encapsulated in a square ceramic boat, having either axial or radial leads. These are generally the lower power types, up to 20 W.

There are two general types of coating applied to wirewound resistors. One is called Pyrosil D-Coat and consists of a combination of silicone resins and refactory material (which prevents oxidation) of the wire element) and is designed for high temperature operation. It is capable of withstanding temperatures corresponding to five times rated load. The other encapsulation material is known as Tropical C-Coat, another silicone compound and is designed to protect the element under environmental conditions (particularly humidity). The power rating is different for similar resistors coated with different coatings. Resistors coated with tropical C-Coat can only operate at half the power of similar resistors encapsulated with Pyrosil D-Coat.

Terminations for wirewound resistors come in a wide variety of styles. The smaller, low power, types (particularly the completely encapsulated types) often have radial or axial leads and sometimes terminal lugs. High power types may have ferrules on each end — and are plugged into large clips; alternatively they may have terminal lugs, Edison screw threads or flying leads.

The resistance element usually consists of nickel — chromium alloy wire (nichrome). Precision wirewound

resistors are usually wound with Eureka wire.

Very high power types and some very low resistance types are sometimes wound with flat-tape element instead of wire. It is usually wound edge-on to the bobbin to improve heat dissipation from the element.

Wirewound resistors are made in

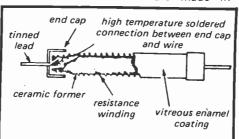


Fig. 1. Typical construction of small, cylindrical style wirewound resistor.

wattage ratings to 250 W, commonly, and up to 1 kW or more for special applications. There are three basic construction styles: cylindrical, flat and encapsulated ceramic-boat style. The first two are also available as adjustable resistors, having portion of the element exposed and a moveable terminal in contact with it.

#### **TEMPERATURE**

Wirewound resistors can have excellent temperature characteristics—as low as 5 ppm/°C, but generally less than 200 ppm/°C for the common types.

These resistors exhibit good stability, usually better than 2%, precision types having stabilities better than 0.05%. Common types are available in tolerances of  $\pm$  5% and  $\pm$  10% depending on

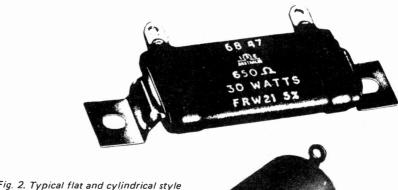
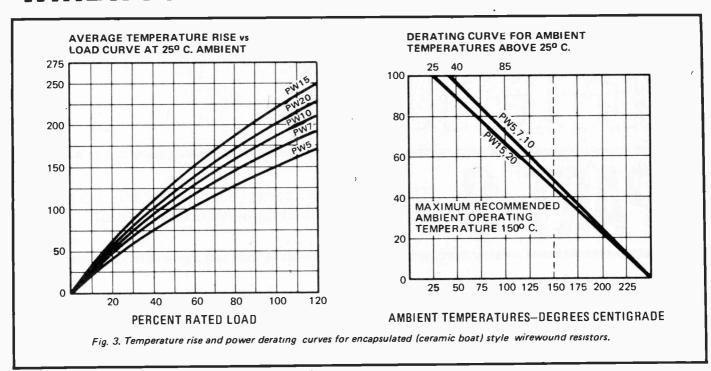


Fig. 2. Typical flat and cylindrical style wirewound resistors.

## **WIREWOUND RESISTORS**



construction style. Tolerance down to 1% can be obtained in precision types.

The noise level and voltage coefficient of wirewound resistors is negligible.

Owing to their construction, wirewound resistors are quite inductive and are generally only useful at low frequencies. Their inherent inductance can be decreased with special winding techniques — occasionally found in precision resistors, but as most wirewound resistors are predominantly used in dc and/or low-frequency circuits where their high power rating is required, this does not present much of a problem.

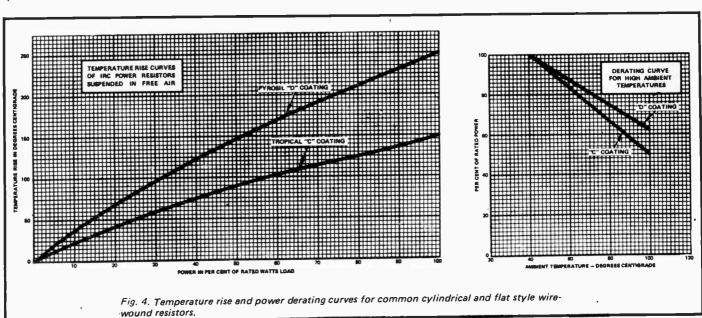
#### **MAXIMUMS**

Wirewound resistors may be operated at temperatures up to 350°C but most common types have a maximum operating temperature (ambient + temperature rise due to power dissipation) of 290-300°C for Pyrosil D-Coat types and 190-200°C for Tropical C-Coat types. Temperature rise and power derating curves for the common cylindrical and flat style resistors are given in Figure 4. The power ratings are based on the ability of the resistor to give long service at full rated load up to the nominated ambient temperature. For higher ambient temperatures, the

resistors are derated according to the curve shown. The full rated load is based on a temperature rise of 250°C from ambient of 40°C for Pyrosil D-Coat and a rise of 150°C from the same ambient for Tropical C-Coat. For the encapsulated lower power varieties, typical temperature rise and derating curves are given in Figure 3. These have a maximum recommended operating temperature of 150°C.

#### **MOUNTING & SURROUNDING**

Care must be taken in the mounting of wirewound resistors to prevent the high operating temperature affecting surrounding components. The cylindri-



cal types usually have a hole through the middle through which heat may escape by convection. Mounting these vertically where possible is recommended to keep their operating temperature down. The flat style are mounted using formed 'leaves' which fit into the ends of the former (see Figure 2) which is hollow, these conducting heat away through the mounting bolts. They are designed for either vertical or horizontal mounting, either singly or in stacks. This style is most suited to applications requiring a high power resistor to be mounted in a limited space. Recommended stacking arrangements are illustrated in Figure 5. When stacked, each resistor affects the temperature of the adjoining resistor(s). To limit the temperature rise of the hottest unit it is necessary to limit the power applied to each resistor (depending on the number of resistors in the stack) according to the percentages shown in the table in Figure 5.

It is a wise precaution with the axial or radial-lead types to mount them so that they are clear of any other components, chassis, pc board, etc by at least their diameter or width, to provide sufficient ventilation and to prevent damage to other components.

#### **FAILURE**

Wirewound resistors fail occasionally. This may be due to one of the following reasons. In high value types, the resistance wire is very thin. The slightest blemish creates a weak point which may eventually cause the wire to break. In the coated types, expansion differences between the ceramic bobbin and the enamel coating may cause cracking of either the coating or the bobbin allowing moisture to penetrate and attack the resistance wire. The wire may corrode under constant dc load con-

TABLE 5. General of	characteristics of	Wirewound	Resistors.
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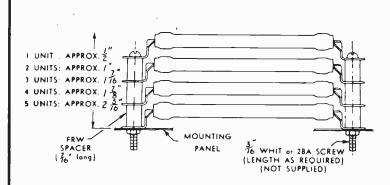
Rated Wattage		Typical Sizes	Typical Resistance Ranges				
(D-Coat)		(Overall)	Fixed Type	Adjustable Type (max.)			
CYLINDR	CAL STYL	E					
(to 40°C)	Length	Diameter	_				
5 W	23 mm	10.3 mm	$0.5 \Omega 5 k$	<del></del>			
10 W	44.5 mm	10.3 mm	$0.75\Omega-12\mathrm{k}$	_			
20 W	50.8 mm	16.7 mm	$1.0~\Omega-25~\mathrm{k}$	5 k			
25 W	63.5 mm	16.7 mm	$1.0~\Omega-30~k$	6 k			
30 W	76.2 mm	16.7 mm	1.5 S2 − 40 k	7.5 k			
40 W	89 mm	23 mm	$3 \Omega - 60 k$	12.5 k			
50 W	114.3 mm	23 mm	$3~\Omega-~88~k$	20 k			
75 W	165 mm	23 mm	$5~\Omega$ $-$ 130 k	25 k			
50 W	81 mm	33.3 mm	$4~\Omega-~80~k$	16 k			
65 W	114.3 mm	33.3 mm	$4~\Omega$ $-$ 120 k	22.5 k			
100 W.	165 mm	33.3 mm	5 $\Omega$ $-$ 200 k	37 k			
150 W	216 mm	33.3 mm	$5~\Omega$ $-$ 270 k	51 k			
200 W	267 mm	33.3 mm	5 $\Omega$ $-$ 340 k	62 k			
FLAT STY	LE (Width	= 14 mm, Mounting	Height = 12.7 mr	m)			
(to 40°C)	Length	Mounting	•				
		Holes $(\phi \text{ to } \phi)$					
20 W	31.8 mm	50.8 mm	$0.5~\Omega - 10~\mathrm{k}$	_			
30 W	50.8 mm	70 mm	$0.5~\Omega-25~\mathrm{k}$	6 k			
50 W	89 mm	108 mm	1.5 $\Omega$ $-$ 50 k	13 k			
65 W	121 mm	140 mm	$2.0~\Omega-20~\mathrm{k}$	19 k			
75 W	153 mm	172 mm	$2.5~\Omega - 100~\mathrm{k}$	25 k			
ENCAPSUL	ATED STY	'LE					
(to 40°C)	Length	Width Height		Inductance (typical)			
5 W	22.2 mm		0.5 - 4.7 k	5.1 μH @ 900 Ω; 20 μH @ 3.3 k			
7 W	35.3 mm		1.0 – 12 k	8 μH @ 2.4 k; 33 μH @ 9 k			
10 W	47.6 mm			13 μH @ 3.9 k; 56 μH @ 15 k			
(to 25°C)			•	το μιτ ε 0.5 κ, 30 μιτ ε 15 κ			
15 W	47.6 mm	12.7 mm 12.7 mm	1.0 - 20 k	13 μH @ 3.9 k; 56 μH @ 15 k			
20 W		12.7 mm 12.7 mm	1.0 - 4.7 k				

ditions due to chemical action in the enamel coating of the component. This latter problem is rare.

Precision wirewound resistors are wound on special bobbins, generally using Manganin wire, and encapsulated or covered in an insulating coating. They are sometimes epoxy-moulded. Other styles are hermetically sealed in a ceramic container. Wire leads or solder lugs are used as terminations. Precision wirewound resistors are not generally designed to dissipate power. Power

types are available however, generally consisting of a conventionally constructed wirewound resistor wound to a tight tolerance or selected, and mounted in an extruded aluminium case. This assists heatsinking, allowing precision resistors to be rated up to powers of 200 W.

The general characteristics of the three basic styles of wirewound resistor are illustrated in Table 5. Typical inductance values for the lower power, encapsulated styles are also given for low and high values.



Kind of "Stack"	% REDUCTION IN RATING When No. of Units Stacked Is					
Mounting Employed	2	3	4			
Resistors Stack-Mounted on a Horizontal Panel Reduce Rating by:	25%	<b>4</b> 5%	60%			
Resistors Stack-Mounted on a Vertical Panel Reduce Rating by:	18%	30%	35%			

Fig. 5. Recommended method of stacking the flat style of wirewound resistors and the necessary power derating. No more than four resistors should be in a stack.



in addition to SCRUMPI Bywood supports SC/MP in these other kits from

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AY5.123	O 7seg. ON and OFF ALARM 5.2	5 TBA

All above clock kits include clock PC board, clock chip, socket and CA3081 driver IC. MH15378 also includes crystal and trimmers. When ordering kit, please use prefix MHI, e.g. MHI 5309

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# BURGLAR ALARM



A simple and versatile control unit, with internal siren and battery.

FOLLOWING ON from last month's Burglarproof Your Home article, we present the ETI burglar alarm. The circuit is simple, reliable and versatile. Based on a single CMOS chip, the standby current is very low, making a mains power supply non essential. Several versions are possible, depending on the particular circumstances where it will be used. We built the simplest version, and will describe it fully, with details of possible modifications and additions.

#### **BASIC UNIT**

The basic unit is self-contained, apart from sensor switches. A 12V HP1 battery is used as the power supply, this battery is capable of powering the system for about a year, if the siren is not

activated! When in the alarm condition the HP1 will power the siren for about 6 hours continously. A battery test facility is included in the design, which displays the on load voltage. We used a Carters "Mini-Mite" 12v siren, which is also inside the case. This siren has a sound output of about 93dB at 3 metres—quite loud!

Other features of the basic unit are, on/off keyswitch on front panel, bell test button and LED fault indication. The box itself is fitted with anti-tamper microswitches, so that it can only be opened in the off position, without sounding the alarm.

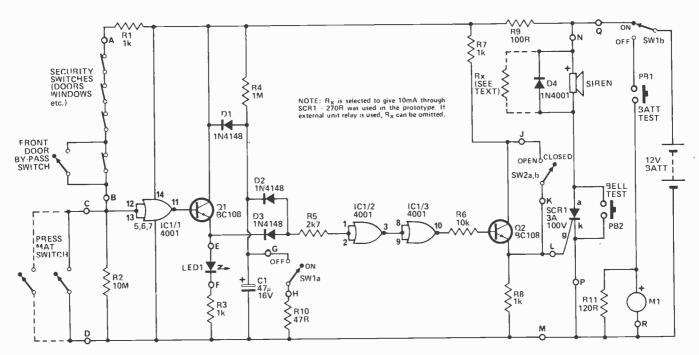
Another important feature is the 30 second delay facility. This ensures that when you switch the unit on — the alarm will not sound for 30 seconds but any

fault will light the front panel LED. Also if you have to walk over a pressure mat or open an alarmed door, to leave, you have the delay to do it in.

#### **SENSORS AND SIRENS**

Three types of sensor can be used with the system. Normally open circuit types, such as pressure mats; and normally closed types such as reed switches biased by magnets. Changeover contacts can also be used, wired to break the normally closed circuit and short the open circuit when operated, this is possible as one wire is common to both circuits. Connection to the unit is via a standard 180 5 pin din plug and socket, all the pins are not used, and external bell

## ETI BURGLAR ALARM



Circuit diagram of basic alarm

units can be wired via this connector if required.

As shown the unit is suitable for shop display protection, caravan protection or even as a tent alarm -- a pressure mat under your ground sheet! For some homes the internal siren will be all that is needed, however, the only way to find out is to try it. If you get lots of complaints from your neighbours or a rapid visit by men dressed in blue -- it's loud enough! Obviously it is best to try it out at a civilised time - not 4 am. If the internal siren is not loud enough, you will need an external alarm. This will be described further on

#### **INSTALLATION**

Bearing in mind the general guide lines given last month, the installation can be worked out. Points to remember are, cover all external doors, and if uneconomical to protect all windows — to cover the internal doors as well, with the odd pressure mat in hallways and on stairs to complete the protection.

Even though single core wire can be used for the closed circuit wiring, we prefer to use 3 or 4 core throughout, as this gives more flexibility in sensors and also creates uncertainty in the mind of a would be by-passer. It also looks like telephone wiring if installed neatly.

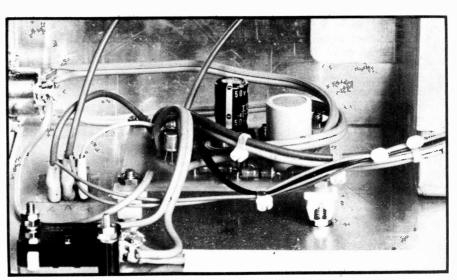
If a shunt switch is not fitted to the main exit, the alarm will sound on entry, this can be unpopular with neighbours. So a shunt switch is strongly recommended.

#### **OPERATION**

In operation the unit is turned on, by the front panel keyswitch. If the LED lights it means that part of the circuit is either open or closed incorrectly, if this happens switch off and find the open door, or chair on pressure mat etc. To check if you have found the fault switch on again

— the LED should remain unlit. A point to note is that you can test the battery with the unit switched off — this is to prevent the battery being flattened by a burglar keeping the test button depressed (if he managed to get up to the box without setting off the alarm).

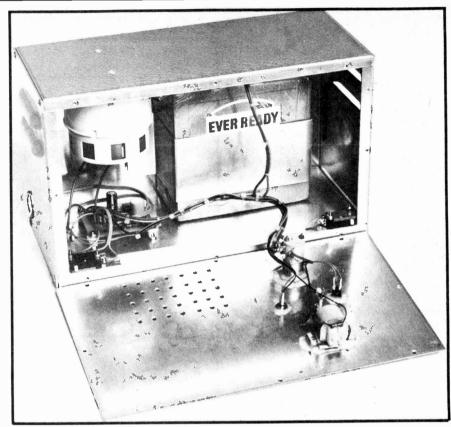
Assuming that you have fitted a shunt switch to the exit, you can now leave and lock the door behind you. This is why a shunt switch inside the main lock is preferred to a separate shunt switch, if you forget to operate a separate switch the door is unprotected.



Close-up of board and wiring, this was an earlier version — hence slightly different to everlay shown

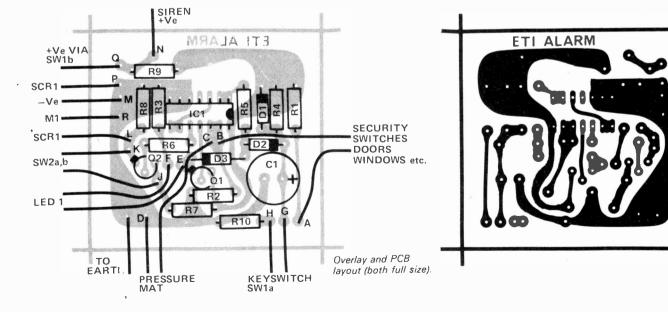
## -How it works-

The circuit is based around a 4001 CMOS quad NOR gate, with the gates connected as invertors (cheaper than using invertors!). The input to ICIa is derived from the closed circuit sensors and the open circuit sensors. In the normal state the output is low and Q1 is off. When either R2 is shorted or the positive supply from R1 is interrupted the gate changes state. Q1 is turned hard on and LED 1 illuminates. R4 and C1 form a timing circuit, which prevents ICIB from giving a low output for 30 seconds from switch on. After this period, if LED1 is on ICIB and ICIC charge state switching on Q2, which triggers SCR1 which self latches. Rx is to make sure SCR1 passes at least 10mA if a bell is used which breaks its own circuit. R11 is selected to draw 100mA to simulate a load for the battery. R10 makes sure that C1 is fully discharged when the alarm is switched off, in order to get consistent timing periods.



General view of the basic unit, note anti-tamper microswitches (SW2 a,b)

Parts L		Semiconductors	i	Miscellaneo	us
		IC1	CD4001A	Meter	Panel type 0-15V
		Q1,2	BC108 or similar		(Doram 259-561)
		D1,2.3	IN4148 or similar	Case	11½ x 5 x 7 inches
Resistors all	5% ½W	D4	IN4001 or similar		(H. L. Smith type W
R1,3,7,8	1 k	SCR1	100V 3A thyristor		£4 59 inclusive)
R2	10M	LED1	TIL 209 or similar	Battery	HP1 + connector
R4	1 M	SW1 (a,b)	Lockswitch		(Sesco 28.63.15 an
R5	2k7		(Doram 337-964)		28.63.55)
R6	10k	SW2 (a,b)	Microswitch	Siren	Carters "Mini-mite"
R9	100R		(normally open)		(Sesco 24.45.20)
R10	47R	PB1,2	Push to make,	PCB, nuts, be	olts, wire, 5 pin Din plug
R11 `	120R		release to break		nsulating kit for SCR1 etc
ŔX	'see text)				, connector and siren
		Sensor switches	see text		m Sesco (Security) Ltd.
Capacitors					d, Hounslow, Middlesex
C1	47 F 16V tantalum			TW3 1TX for	£9.23 inclusive.



## ETI BURGLAR ALARM

On returning open the main door and then switch the unit off.

#### CONSTRUCTION

Construction is quite straightforward, most of the components are mounted on a PCB. The main point to watch is that CMOS is involved, the usual precautions should be taken. Make sure your iron is earthed and fit the IC last. All bolts must be fitted with two nuts to prevent external removal. The front panel mounted parts should be epoxyed into place, also to prevent external tamper-The two microswitches SW2a+b should be fitted so that the front panel keeps them depressed when in place. General layout is easily seen from the photographs.

The unit should be screwed in position through the back panel when complete.

#### **EXTERNAL UNITS**

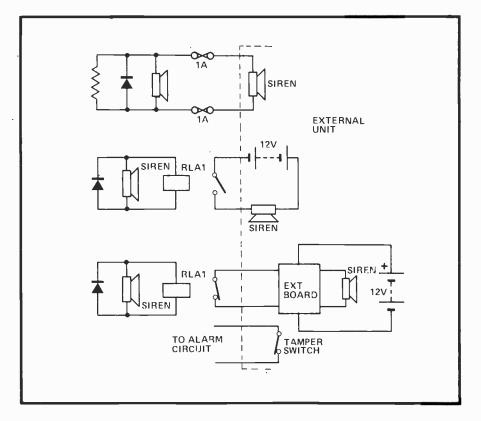
If an external siren or bell is needed there are three ways this can be done. The simplest, but least secure is to run it in parallel to the internal siren via twin wires. A 1 A fuse should be placed in each of the leads — so that a short circuit will not flatten the battery. Obviously the battery life will be reduced when powering two alarm sirens.

The second method is to run a relay in parallel with the internal siren, the external unit then needs its own power supply. The relay should be mounted at the main unit. The external power supply can be another battery or a mains power supply.

The disadvantage of both these methods, is that if the wires are cut the external alarm will not sound. Although if the cable is run inside metal tubing this is not so much of a problem.

#### **BEST METHOD**

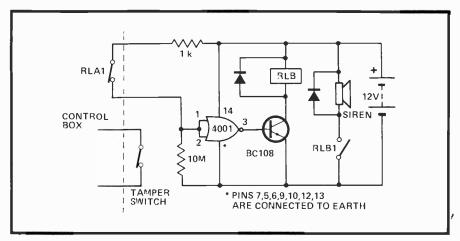
The best method is to use a sensor circuit with a relay output. This can be obtained from another CMOS circuit similar to the main unit — only simpler. Again a relay is connected across the siren but if the wires to it are cut the external siren will sound. The standby current is about 1—A so the battery can be left connected permanently. A suggested interwiring diagram is shown.



Top: Simplest external siren circuit.

Middle: Addition of a 12 volt relay (RLA1) gives another simple external siren circuit.

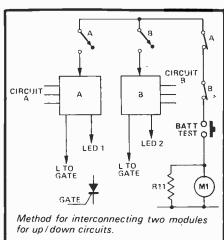
Bottom: Most secure method, involves the use of a CMOS chip, circuit below.



#### **UP AND DOWN**

Another possible modification is to add an upstairs/downstairs facility to the main unit. This can be done by duplicating the main board and adding an extra keyswitch to the front panel, together with a second LED. The interconnection for this is also shown.

The batteries should be checked at least once a month, and replaced when on load they register below 11 volts.



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ł	CD4000	0.17	CD4028	1.31	CD4054	1.33	CD4095	1,20
Ì	CD4001	0.18	CD4029	0.64	CD4055	1.51	CD4096	1.20
I	CD4002		CD4030	2.55	CD4056	1.51	CD4097	4.28
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1	CD4013	0.64	CD4038	1.24	CD4069		CD4516	1.56
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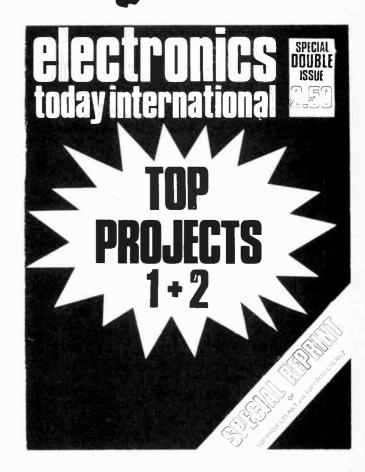
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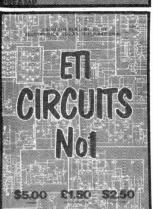
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APPLICATIONS: Hi-Fi — Disco — Monitor — Power Slave — Industrial — Public address.

SPECIFICATIONS:

120 Watts into  $8\Omega$ 

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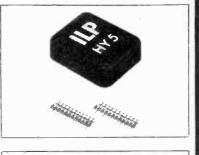
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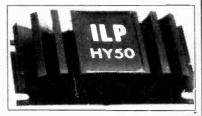
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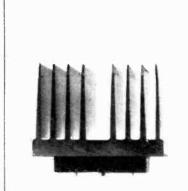
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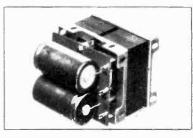
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# ELECTRONICS —it's easy!

#### TRANSMISSION LINE THEORY

PART 38

ELECTRONIC systems consist of basic analogue and digital subsystems interconnected to provide the required overall input-output relationships. It is important for the various subsystems to be interfaced correctly if they are to perform as intended. But with this condition satisfied, one cannot just assume that subsystems merely connect together without need to consider any other parameters in the interconnection process.

In practice the individual circuit assemblies may be geographically apart — such as the remote control of off-shore oil wells by a shore-based computer, the recording of test data from a missile, the control of banking accounts by a central computer centre or the sensors of a refinery which connect to the central control room. Each of these required some form of telemetry system.

When making connections it is also important, especially when noise sources are present that will interfere with the signal, to ensure that the signal is transferred from stage to stage without significant noise pick-up or signal degradation.

#### TRANSMISSION LINKS

Several different transmission methods exist in which the signal is confined - open wires, coaxial cables and waveguides, optical fibres etc. Alternatively, information can be transferred via open radiation paths - radio, optical or acoustic links. The required signal bandwidth is one of the primary factors deciding which method is used. In radiation methods it is often necessary to use a carrier frequency higher than the signal bandwidth dictates because low frequency carriers will not radiate as well for the same amount of transmitted power.

Confined Signal Links: The simplest links are formed using an open-wire circuit (supported on insulators) or a multicore cable (such as is used in local telephone distribution).

Although apparently trivial, lines may, in fact, be an important part of the system. They are not as simple as they first appear because they have a frequency response that must be adequate for the signal bandwidth to be transmitted. Open-wire lines would not normally be used beyond 10MHz. Above that coaxial cables are needed — these are useful to about 5000MHz.

When current flow in a conducting line, magnetic and electric fields are set up around the wires. Figure 1 shows these plotted for the various kinds of cable. Open configurations radiate energy, the amount increasing with the fre-

quency of the signal. A line is, in reality, a distributed inductance and capacitance component which also has losses due to the resistance of the wire and the resistance to ground. Figure 2 shows how lines can be considered as a lumped-element equivalent circuit which can be analysed more easily. Depending upon the factors that are negligible for a particular case the equivalent can be reduced to simpler circuits see Fig. 3. For example, at very low frequencies (less than say 100 kHz) a medium length line may be represented by the series resistance of the cable shunted by the capacitance of the line. Typical

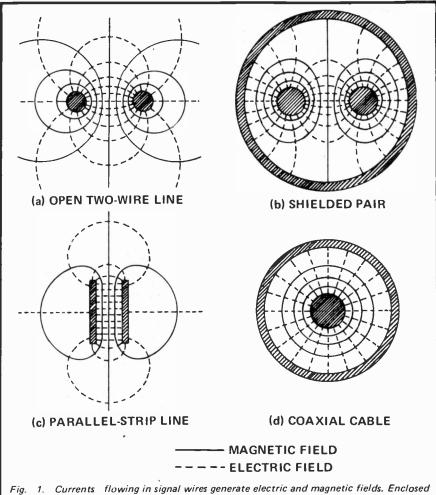


Fig. 1. Currents flowing in signal wires generate electric and magnetic fields. Enclosed configurations can be used at higher frequencies because these fields are contained.

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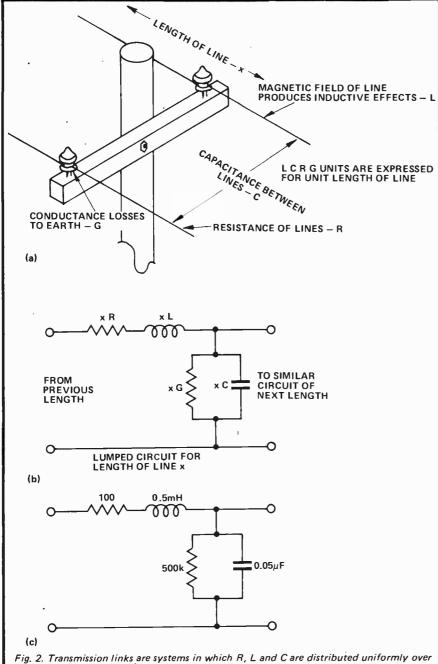


Fig. 2. Transmission links are systems in which R, L and C are distributed uniformly ove the length. For convenience we can consider the line as being composed of cascaded lumped-equivalent elements.

- (a) A length of low frequency telephone line.
- (b) Approximate lumped-element equivalent.
- (c) Representation values for 1 km. of medium-size telephone line with earth return.

(Actual constants vary widely depending upon design of line).

cables may have a resistance of around 0.05 ohm per metre and a capacitance of 100 pF per metre. Hence a long length of shielded or open cable could provide a considerable shunting effect that attenuates and phase shifts the signal.

#### **APPLY OHMS LAW**

When connecting high outputimpedance sensors to lines, as little as one metre of cable may be sufficient to markedly attenuate the signal. It's a matter of applying Ohms law to the suitable equivalent circuit.

Because of the reactive effects of the cable the higher frequency signals transmitted will be degraded more than the low frequencies — for example, square waves become rounded as well as attenuated. The high-frequency performance of the line may be improved by "loading" it with inductors placed at regular intervals. The inductance value is

chosen to tune out the inherent capacitive reactance at the upper frequency where response begins to fall off, a method that extends the bandwidth some way beyond the inherent, unloaded upper limit. This is used, for example, to broaden the bandwidth of submarine cables.

#### **CO-AXIAL**

The coaxial cable, shown in Fig. 4, by virtue of the surrounding external shield (Fig. 1) acting as the second wire, has no external field and, therefore, does not radiate energy. Because of this a well designed coaxial cable will pass from dc to microwave frequencies that is, such a cable can have a bandwidth of about 5000 MHz. Coaxial cable is, therefore, potentially able to transfer much more information than open wires. It does however need a common earth connection (asymmetric) and can't be used in a balanced mode (see later). The bandwidth of practical coaxial cables is limited by resistive and dielectric losses. In practice waveguides are generally used at frequencies above 1000 MHz or so.

#### **WAVE GUIDES**

Waveguides consist of precise pipework — they look as if they had been made by a precision plumber! Waveguides carry travelling electromagnetic waves of very high frequency and behave vaguely in the same way that pipes carry water. They cannot however be used for low frequency transmission.

The cross-sectional area of a waveguide is inversely proportional to the design frequency. As a general rule of thumb guide the upper frequency limit of a waveguide is where the wavelength of the signal becomes one quarter of the guide aperture — millimetre wavelength signals (50 GHz or so) being the practical upper limit.

#### **OPTICAL FIBRES**

Beyond this, a still wider bandwidth is obtainable using optical fibre transmission elements which will pass radiation in the visible light region) 10<sup>14</sup> Hz to 10<sup>15</sup> Hz). At our current state of technology, however, scientists have only been able to detect the frequencies of far infra-red signals (around 10<sup>11</sup> Hz). We cannot, as yet, monitor individual cycles of light with electronic detectors.

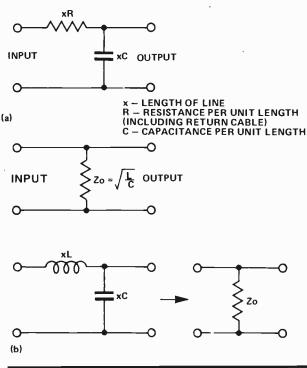
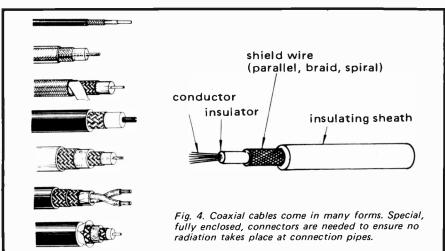


Fig. 3. In certain practical cases the lumped equivalent reduces to simpler situations situations.

(a)

Low frequency (negligible L assumption) short line in which only C and R are dominant. The R, C values are found from maker's data.

(b) High frequency lossless line (negligible R and G assumptions). The input and output impedances of the line are equal and constant regardless of length.



#### **LUMPED LINES**

When the losses of the line are insignificant (G = 0, R = 0, in Fig. 2b) the lumped-equivalent of the transmission lines reduces to L in series and C shunting, as shown in Fig. 3b. The nett result is, rather surprisingly, that the line exhibits only resistance of a fixed value when looking into the ends. This is called the characteristic impedance,  $\hat{Z}_{0}$ , for which  $Z_{0} = (inductance per$ unit-length/capacitance per unit length)  $\frac{1}{2}$ . The line appears to be purely resistive and the Zo value is decided by the design of the line or cable, not by its length! Examples are 600 ohm telephone lines, 75 ohm colour TV coaxial feeder cable. This means, in practice, that we can interconnect units on the basis of matching all connections to the Zo of the cable without having to worry about the cable length. If this rule is observed, no high-frequency energy

will be reflected at the termination to change the information being transmitted. (The need for correct matching was also mentioned in the previous discussion about filters). However, if the line is very long matching must still be applied to obtain maximum transfer, but account must now be taken of losses. For example a typical 75 ohm coaxial cable will have losses of the order of 2 to 5 dB per one hundred metres.

Radiation Links: Electrical signals fed into open wires radiate energy out into the surrounding medium. As well as this radiated energy there also exists a 'near field' that remains established, storing energy. This is the field we associate with, say, an electromagnet. As the frequency rises the ratio of radiated energy to stored energy increases. For this reason we are

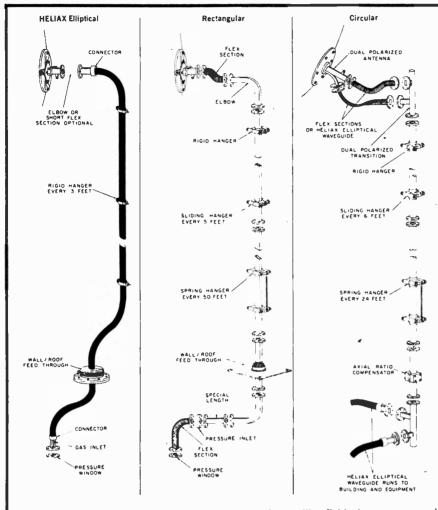
able to build efficient radio systems provided the frequency is kept above 100 kHz or so. Lower frequencies can be used as transmission systems but the power input needs rise enormously for the same distance radiated in free space. (The Omega navigation system uses extremely powerful VLF signals because of their ability to penetrate deep into the waters of the ocean). Beyond the gigahertz frequency region, circuitry becomes impracticable with current technology.

Even though the radiated energy must be at a very high frequency to operate efficiently we may not necessarily need to use the bandwidth available on the carrier, modulation techniques are used to super-impose a relatively narrow bandwidth signal on to the carrier. It might be thought that optical and infra-red links use extremely high carrier frequencies (330000 GHz for red light) but in these applications the carrier is not modulated on an individual cycle basis but rather as a variation of a continuous dc link. Fig. 6 is a modern link designed to transmit television plus speech a bandwidth of commands 7.5MHz. Acoustic links using soundwave propagation operate with frequencies as low as 10Hz to well above the 10MHz region. These can be modulated on the individual cycle basis.

Skin Effect: The alternating magnetic field produced around a wire has the effect of causing the current flowing in the wire to flow at a greater density in the outer region of the wire. The higher the frequency the more pronounced this so-called skin-effect. At the very high frequencies so little current flows in the centre of the cable that the centre is often omitted completely, thus a tube is used as a conductor. For example, at 1 MHZ the majority of the current flows in a copper cable to a deoth of only 60 um whereas at 60 Hz the distance would be 8.6 mm depth. This also means that the effectivle resistance of a wire rises significantly with frequency — by factors of 100.

Process Industry Telemetry Links: Process plants such as oil refineries, paper mills, brick kilns, power stations and aluminium refining plants are monitored by using hundreds of sensors connected to the control-room area via instrumentation links. These are invariably wired using shielded wire or coaxial cable. Because of the extreme electrical noise level of

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 ${\it Fig.~5.~Waveguides~transmit~electro-magnetic~energy~via~travelling~fields~that~are~conveyed~through~the~pipework.}$ 

such plants and low output signal level of the senors these links could pick up significant noise thus degrading the senor information. Over the years process instrument suppliers have standardised the design of the control systems, and their installation and noise pick-up by the cable has been avoided by several methods.

The first strategy is to superimpose the information signal on to a standing current or voltage thus raising the wanted signal level above expected noise levels. The two systems commonly used transmit the signal range of the data through 4-20 mA DC or 10-50 mV DC systems. An 0-20 mA system is also common. Current transmission has the advantage that the circuit is of low impedance - a few ohms which reduces the level of induced noise power. Figure 7 is an example of these practices - Honeywell's arrangements used to test the temperature and pressure of natural

gas wells in the Leman Field of the North Sea.

#### **SAFETY PRECAUTIONS**

Often the sensor has to be placed at a location where an explosion could result from a spark or excessive overheating of a malfunctioning sensor circuit. The most obvious way of overcoming this is to place the whole unit in an explosion-proof enclosure. This, however, has disadvantages the cost is high, and testing and maintenance difficult due to the need to shut off the power when the enclosure is opened.

The alternative, more modern, method is known as intrinsic safety. As inflammables require a specific level of energy to ignite them, explosion can be prevented by ensuring that the sensor stage cannot, under any conditions, provide enough ignition energy. No enclosures are needed and the circuit can be maintained whilst it is operating. Originally the concept

was implemented by ensuring the senor circuitry could not draw, or produce via storage, more than a specified power level. This level was found by experiment in a test rig set up for the situation involved.

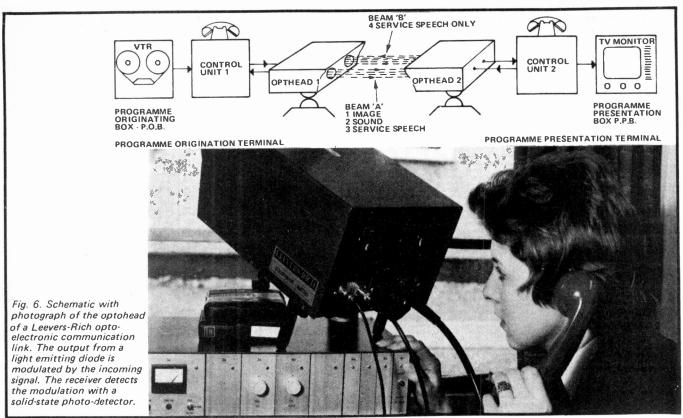
The more recent idea is to use "safety barriers." At the exit from the declared hazardous area, the cables terminate into a zener-diode and attenuator arrangement which ensures that the current and voltage entering the area are limited to safe values. Figure 8 shows the circuit of a zener barrier. Another safety device uses a solid-state closelycoupled electro-optic link which provides DC electrical isolation between its input and output, the information being transferred from a light-emitting-diode mounted next to a silicon photo-diode detector. These ensure that overvoltage or induced earth-loop currents cannot enter the isolated hazardous area.

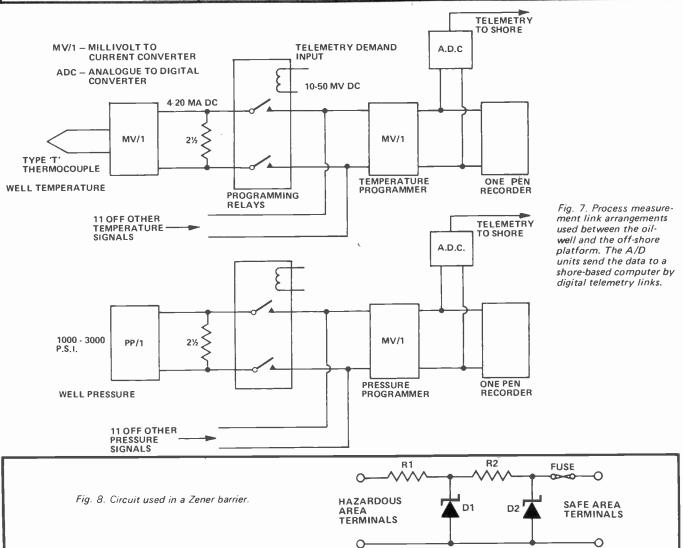
#### **MEDICAL MATTERS**

In electro-medical instrumentation, safety precautions of another kind are vital to ensure the senor does not act as a pathway for a dangerous level of electric-current into the patient. At 240 VAC the human body's resistance, hand to hand is around 2000 ohms - 100 mA will flow. If totally connected (as by a conducting fluid) the resistance reduces to 200 ohm - 1 A will flow. About 75mA through the body will produce heart fibrillation; only 150 aA, through the heart itself, is needed to produce this effect. A person can usually hold (with the fingers) and release as much as a 10mA, 240 VAC current - beyond that the muscles become paralysed. Skin moisture largely decides the hand to hand resistance. When dry it will be (at 240 V) ohms and moist 1000 ohms. Thus a hand-to-hand 240 V encounter will provide a shock at least double the fibrillation level!

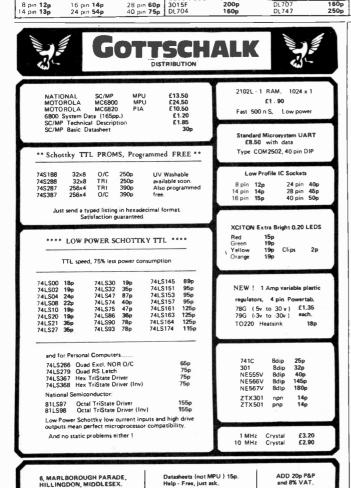
The instrumentation must, where the metal parts are earthed, be wired with the active, neutral and earthing wires connected correctly. Double-insulated systems avoid this problem. Earth-leakage balanced — core breakers are worth using. These defect minute difference currents in the active and neutral, tripping a breaker if they rise above milliamperes.

The sensor attached to the patient must not be capable of providing a lethal level of energy by means of feedback from the instrumentation.





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	18p	74107	36p	4007	21p	CA3028A	Diff Cascade Am		099	112p	AF115	22p	BFX84	30p 30p	2N918	43p	40411	325p	2.7 to 33\	,	6A 200V 108p
	96p	74109	96p	4009	67p	CA3046	5 Transistor Arra		4 pin DIL	85p	AF116 AF117	22p 22p	BFX85 BFX86	30p	2N930 2N1131	19p 20p	40594	90p	400mW	11p	6A 400V 120p
	14p	74110	55p	4011	21p	CA3048	4 Lo Noise Amp		6 pin DIL	250p	AF139	43p	BFX87	30p	2N1131	20p	40595	97p	100	22p	TRIACS
	lOp  Sp	74116	216p	4012	19p	CA3053	Diff. Cascade Am		O5/DIL	70p	AF239	48p	BFX88	30p	2N1304	45p	i .		l '**		Plastic
	13p	74118 74120		4013 4015	55p	CA3080	Op. Transcond A	Amp. 8	DIL	97p	BC107/B	10p	BFY50	18p	2N1305	45p	FETs				Amp Volts
	24o	74121	32p	4015	90p 54p	CA3089E	FM IF System		6 pin DIL	250p	8C108/B		BFY51	16p	2N1306	48p	BF244 MPF102	36p 40p	1		3 400 85p
	10p	74122	52p	4017	110p	CA3090A0			)IL	500p	8C109/C		8FY52	18p	2N1613	27p	MPF 102	40p	}		6 400 107p
	33p	74123	73p	4018	247p	ICT8038CC			6 pin DIL	370p	8C147	90	8RY39	45p	2N1711	27p	MPF103	40p	NOISE		6 500 <b>120p</b>
	IOp	74126	75p	4020	140p	LM380N	2W Audio Amp		4 pin DIL	115p	BC148	90	BSX19	20p	2N1893	32p	MPF105	40p	75J	140p	10 400 <b>140</b> p
	39p	74132	75p	4022	180p	LM381N	Stereo Pre Amp		4 pin DIL	190p	BC149	10p	BSX20	20p	2N2219	25p	2N3819	27p			10 500 160p
	18p	74136	81p	4023	19p	LM389N	Aud Amp +3 Ti		8 pin DIL	175p	BC157	11p	BU105	175p	2N2222	25p	2N3820	50p	1		15 400 <b>200</b> p
	37p	74141	80p	4024	100p	M252	Rhythm Generate		6 pin DIL	850p	BC158	13p	BU108	312p	2N2369	15p	2N3823	54p	DIAC		15 500 <b>225p</b>
7437 3	37p	74142	300p	4025	19p	MC1310P MC1351P	FM Stereo Decode Lim/Det. Aud P		4 pin DIL 4 pin DIL	190p 104p	BC159	13p	MJE340	49p	2N2484	32p	2N5457	40p	BR100	30p	40430 <b>130</b> p
	37p	74145		4026	200p	MC3340P	Electronic Attenu		3 pin DIL	180p	BC169C	15p	MJ2955		2N2904/		2N5458	40p			40669 <b>130p</b>
	l 8p	74148		4027	81p	MFC4000E		Halloi C	CB DIL	90p	BC171	12p	MJE2955		2N2905/		2N5459	40p	MEMOR	Υ	
	35p	7415D		4028	152p	NE54OL	Aud Pwr. Driver		05	175p	BC172	12p	MJE3055 MPSA06		2N2906	25p	3N128	95p	2102 RA		€2.70
	75p	74151		4029	130p	NE555V	Timer		pin DIL	40p	8C173	13p 20p	MPSAUD MPSA12		2N2026F 2N29260		3N140	95p	2107 RA		£10.80
	16p	74153		4030	59p	NE556	Dual 555		4 pin DIL	960	BC177 8C178	17p	MPSA56		2N29260 2N3053	20p	3N141	95p	2112 RA	MA	£4.70
7444 11		74154		4042	150p	NE5618	PLL with AM Dei		6 pin DIL	425p	BC178	20p	MPSU05		2N3053	54p	40603	63p	2513 RO		£8.50
	8p	74155	96p	4043	218p	NE5628	PLL with VCO		6 pin DIL	425p	BC182	12p	MPSU06		2N3055	54p	40673	70p	745262	N RO	M £24.50
	9Op	74156 74157	96p 97p	4046 4047	150p 110p	NE565	PLL	1	4 pin DIL	200p	BC183	12p	MPSU55		2N3442	151p	UJTS				
	85o	74160		4047	68p	NE566V	PLL Fun Gen.		pin DIL	200p	BC184	14p	.MPSU56		2N3702	14p	TIS43	40p			
	20p	74161		4050	50p	NE567V	PLL Tone Decode		3 pin DIL	200p	BC187	32p	OC28	90p	2N3703	14p	2N2160	95p			
	20p	74162		4054	130p	2567	Dual 567		6 pin DIL	400p	BC212	14p	OC35	90p	2N3704	14p	2N2646	48p	SCR THY		
	20p	74163		4055	140p	SG3402N	Ring Modulator		14 pin DIL	275p	BC213	12p	.0C71	25p	2N3705	14p	2N4871	65p	1A 50V		43p 45p
7460 2	20p	74164		4056	145p	SN72710	Diff Comparator		4 pin DIL	54p	BC214	17p	TIP29A	50p	2N3706	14p			1A 100V		43p 50p
747D 3	32p	74166	136p	406D	130p	SN72733	Video Amp		14 pin DIL	150p	BC478	32p	TIP29C	62p	2N3708	14p	PUJT		1A 400V 3A 400V		81p
7472 3	12p	74167	370p	4069	30p	SN76003N	N Aud Pwr Amp. 10W Amp, in 4		l 6 pin DIL 5 pin Plastic	275p 280p	BC547	12p	TIP30A	60p	2N3709	14p	2N6027	60p	8A 50V		142p
	16p	74174		4071	29p	SN76008 SN76013			opin riastic 16 pin DIL	175p	BC557	12p	TIP30C	72p	2N3707	14p	DIODES		16A 400V		220p
	36p	74175		4072	29p	SN760231			16 pin DIL	175p	BCY70	22p	TIP31A	56p	2N3773	270p	SIGNAL		16A 600V		270p
	48p	74176		4081	21p	SN760231			16 pin DIL	275p	BCY71	24p	TIP31C	68p	2N3B66	97p	OA47	10p			2.00
	37p	74177		4082	29p	TAA621A	Aud Amp. for T		ΩIL	270p	BD124 BD131	140p 39p	TIP32A TIP32C	63p 85p	2N3904 2N3905	22p 25p	0A47	15p	BT106 1A	700V	STUD 130p
	54p	74180		4510	142p	TAA6618	FM/IF Amp Lin		ail.	150p	BD132	43p	TIP33A	97p	2N3905 2N3906	22p	0A85	15p	C106D 4		
	08p	74181		4511	200p	TBA6418	Audio Amp		ΣIL	300p	BD135	54p	TIP33C	120p	2N4058	19p	0A90	9 p	MCR101		V TO92 27p
	88b   83b	74182 74185		4516 4518	140p	TBA800	5W Audio Amp.		ΩIL	100p	BD136	55p	TIP34A	124p	2N4060	19p	OA91	9p	2N3525		
	03b	74185		4518	140p	TBA810	7W Audio Amp		ΔIL	125p	BD139	54p	TIP34C	160p	2N4123	22p	OA95	9p	2N4444 8		/ Plastic 200p
7/484 10	osp	74130	1000	4520	1300	TBA820	2W Audio Amp		JIL .	100p	BD140	60p	TIP35A	243p	2N4124	22p	OA200	8р	2N5060		V TO92 36p
VOLTAGE	REGI	ULATOR	IS			TDA2020	20W Audio Amp		OIL OIL	375p	80Y56	225p	TIP35C	290p	2N4125	22p	OA202	10p	2N5062		
Fixed-Plastic		rminals				XR2240	Prog Timer/Coi		16 pin DIL		BF115	24p	TIP36A	297p	2N4126	22p	1N914	4р	2N5064	3 8A 2C	OV TO92 43p
1 Amp +v	ve			-ve		ZN414	TRF Radio Recei		TO18	140p	BF167	25p	TIP36C	360p	2N4401	34p	1N916	11p			
5V 7805		150p		905	215p	OPTO-FILE	ECTRONICS	DRIV			BF173	27p	TIP41A	70p	2N4403	34p	1N4148	4р	L		
12V 7812		150p		912	215p		RANSISTORS	7549		84p			11011/5	DE:	CEC	A 4.3	20- D	e n		hor	0 W \$ # 0 0
15V 7812		150p		915 918	215p 215p	OCP70	40p	7549	2	104p									– no ot		
18V 7819		150p 150p		918	215p	OCP71	120p	LEDS	0.0-4	10-	BAALL	OPI	SER OF	NI V	GOV	TC	OLLEG	ES I	ORDER	SW	ELCOME
24V 7824 LM309K	4 5V	1 Am		TO3	150p	2N5777	50p		9 Red	16p 36p	LIAIWIF	UNI	JEN UI	WE .	300		OLLEG		UIIDEN	2 44	LLOOIVIE
LM309K	5V	100m		105	97p	LDRs	· ·		1 Green Infrared	81p							A 7				TD
	12V	0.54		TO5	106p	ORP12	70p	0.2	imareo	a ib			- 1				//				
VARIABLE		,-				ORP60	75p	Red		18p				1							
	4pm [	DIL			45p	ORP61	75p	Green	,	36p			, , ,								ΓD
			C CV =	EVAC.				Volto:		32p									Tel. 01		
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# Unique full-function 8-digit wrist calculator... available <u>only</u> as a kit.

A wrist calculator is the ultimate in common-sense portable calculating power. Even a pocket calculator goes where your pocket goes – take your jacket off, and you're lost!

But a wrist-calculator is only worth having if it offers a genuinely comprehensive range of functions, with a full-size 8-digit display.

This one does. What's more, because it is a kit, supplied direct from the manufacturer, it costs only a very reasonable £9.95 (plus 8% VAT, P&P). And for that, you get not only a high-calibre calculator, but the fascination of building it yourself.

#### How to make 10 keys do the work of 27

The Sinclair Instrument wrist calculator offers the full range of arithmetic functions. It uses normal algebraic logic ('enter it as you write it'). But in addition, it offers a % key; plus the convenience functions  $\sqrt{x}$ , 1/x,  $x^2$ ; plus a full 5-function memory.

All this, from just 10 keys! The secret? An ingenious, simple three-position switch. It works like this.



1. The switch in its normal, central position. With the switch centred, numbers – which make up the vast majority of key-strokes – are tapped in the normal way

2. Hold the switch to the left to use the functions to the left above the keys...

3. and hold it to the right to use the functions to the right above the keys.

The display uses 8 full-size red LED digits, and the calculator runs on readily-available hearing-aid batteries to give weeks of normal use.



3

Dimensions: 113/16" (46 mm) wide 17/16" (37 mm) deep. Weight: less than 1 oz (28 g).

KIT ONLY £9.95 PLUS VAT, P&P

Sinclair Instrument Ltd, 6 Kings Parade, Cambridge, Cambs., CB2 1SN. Tel: Cambridge (0223) 311488. Assembling the Sinclair Instrument wrist calculator

The wrist calculator kit comes to you complete and ready for assembly. All you need is a reasonable degree of skill with a fine-point soldering iron. It takes about three hours to assemble. If anything goes wrong, Sinclair Instrument will replace any damaged components free: we want you to enjoy assembling the kit, and to end up with a valuable and useful

ESUSETE

calculator

Contents
Case and display
window.
Strap.
Printed circuit board.

Switches.
Special direct-drive chip
(no interface chip needed).
Display.

Batteries

Everything is packaged in a neat plastic box, and is accompanied by full instructions. The only thing you need is a fine-point soldering iron.

All components are fully guaranteed, and any which are damaged during assembly will be replaced free.

The wrist-calculator kit is available only direct from Sinclair Instrument. Take advantage of this 10-day money-back undertaking.

Send the coupon today.

#### To: Sinclair Instrument Ltd, 6 Kings Parade, Cambridge, Cambs., CB2 1SN.

- \* Please send me... (qty) Sinclair Instrument wrist-calculator kits at £9.95 plus 80p VAT plus 25p P&P (Total £11).
- \* I enclose cheque/PO/money order for £.....
- \* Complete as applicable.

Name .

Address

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I understand that you will refund my money in full if I return the kit undamaged within 10 days of receipt.

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74C08	0.25	74C90 0.90	74C1751.21	74C9107.18
74C10	0.25	74C93 0.90	74C1921.48	74C9141.51
74C14	1.51	74C95 1.31	74C1931.48	74C9182.89
74C20	02.5	74C1071.29	74C1951.31	74C9258.28
74C30	0.25	74C1512.63	74C2007.19	74C9268.28
74C32	0.25	74C1543.92	74C2211.49	74C9278.281
74C42	1.93	74C1572.35	74C9010.72	74C9288.28
74C48	2.37	74C1601.48	74C9020.72	80C95 0.72
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0.60	4034					1.16
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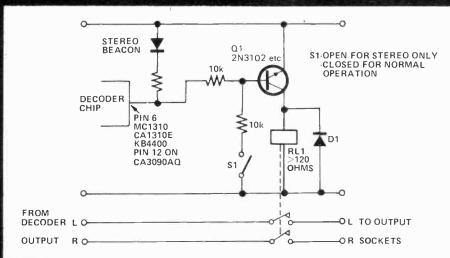
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ETI is prepared to consider circuits or ideas submitted by readers for this page. All items used will be paid for. Drawings should be as clear as possible and the text should preferably be typed. Circuits must not be subject to copyright. Items for consideration should be sent to ETI TECH-TIPS, Electronics Today International, 25-27 Oxford St., London W1R 1RF.



#### STEREO ONLY

This circuit allows only stereo broadcasts to be outputed by a tuner using either a 1310 or 3090 type stereo decoder chip. In both cases the stereo beacon driver is used to switch the audio output of the tuner. When a stereo signal is being received the beacon driver output is low which turns the Q1 and energises reed relay RL1. The two contacts which switch the output lines are closed and the stereo signal is available at the tuner output sockets. RL1 can be any reed relay with a coil resistance greater than 120 ohms and two normally open contacts.

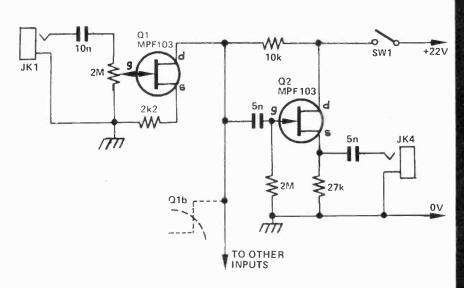
#### **BASIC MIXER**

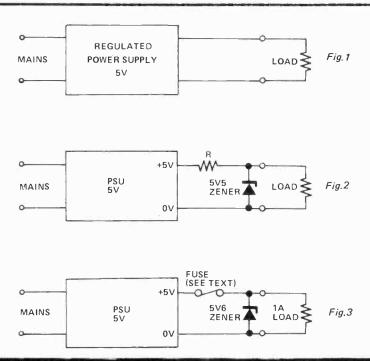
This simple mixer circuit will work with two or three channels, providing excellent input isolation and exceptional frequency response, extending well over the top end of the audio spectrum.

It is usable by one or more instruments plus microphone, or with special effects, such as mixing an input with pink noise, to give 'surf'.

The unit will give 8db gain, and since low-level signals are involved, should be housed in an aluminium box. If a mains supply is used, the usual anti-hum precautions must be taken.

It is useful to use scaled slider potentiometers, so that effects may be re-created.





#### PROTECTION FROM TTL PSU FAILURE

With this circuit, a fault in the sophisticated PSU might cause the output voltage to rise above about 5.5V, (the maximum allowable) and thus cause damage to the ICs.

A simple zener regulator across the output as in Fig. 2 with a zener voltage of about 5.5V, means that at normal voltage, the zener is effectively open circuit. The effect of the load resistor R, would be to eliminate all the regulation of the main PSU.

In the circuit shown in Fig. 3, there is no load resistor to cause regulation problems, and the zener normally appears as an open circuit. But as soon as the voltage rises above about 5.5V the zener tries to draw a great deal of current and the fuse blows, cutting off the supply from the load.

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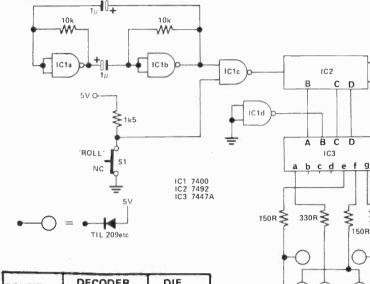
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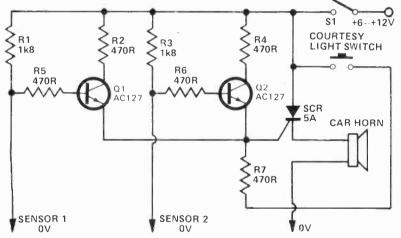
COUNT	DECODER INPUT	DIE DISPLAY
0	2	5
1	3	3
2	6	6
3	7	1 !
4	10	4
5	1 1	1 6

#### 3 CHIP DIE

This differs from previously published circuits in that decoding, count and drive LED is achieved by a single 7 segment decoder/driver chip.

IC1a and b form a multivibrator, providing clock pulses for the counter IC2. IC1c gates the pulses to the counter when the 'roll' switch, S1, is opened. IC1d is used to provide a logic 1 for the B input of the decoder, IC3

**≥** 150R



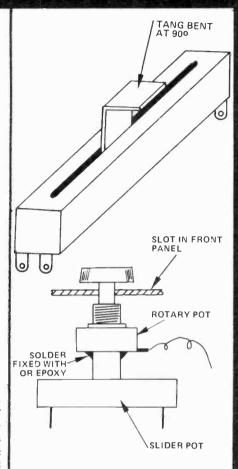
#### CAR RADIO PROTECTOR

Many circuits have appeared for protecting radios and stereos in soft-top motor cars whose interiors are readily accessible to thieves. These circuits however, have the disadvantages of high parts count and expensive relays to switch on and latch the alarm.

The circuit operates as follows: sensor leads 1 and 2 are connected to the chassis of the equipment to be protected, therefore holding the bases of Q1 and Q2 at earth potential, and

thus switched off. If one of the sensor leads is broken, current flows to the base of the respective transistor and switches it on. This gates the SCR and sounds the alarm. The self-latching characteristics of the SCR now make the transistor and its sensor lead inoperative and the alarm can only be stopped by switching off the concealed switch, S1.

To prevent thieves tampering with the wiring, a seperate car horn and courtesy light switch (obtained from a breaker's yard) were fitted under the



#### **JOYSTICK**

Shown is an idea used successfully to provide a 'joystick' type of control with a television football game, by mounting an ordinary rotary type potentiometer on the tang of a slider potentiometer. The rotary control is attached to the circuit board via flying leads.

Radio control enthusiasts could use the idea in conjunction with a proportional system, giving a very cheap alternative to 'joysticks'.

bonnet. If thieves cut the normal horn wires the alarm is unaffected; also any attempt to lift the bonnet to disconnect the battery will trigger it. The sensor leads are multistrand flexible cable with only one strand connected to the equipment, therefore easily broken while trying to remove it from the dashboard.

The transistors used are not critical and most NPN. general purpose transistors should suffice. The stand-by current is very low (typ. 13mA), and therefore is designed to be left switched on. The owner can never forget to switch on the alarm when leaving the vehicle. He must, however, switch off before lifting the bonnet.

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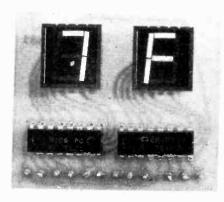
Hobby

- entertainment

- games

# tech-tips

# DISPLAY



**FND 500** ٥v c d e f 12 11 10 9 15 14 16 5 9368 STROBE BINARY INPUTS

The circuit diagram of one half of the display.

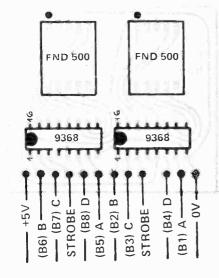
THERE ARE MANY SITUATIONS in which the computer enthusiast wants to see the data on a parallel binary data bus (such as the one carrying the output from the terminal keyboard in this issue). Certain conventions have arisen to provide standard ways of displaying and manipulating large binary data words - because we are not equipped to handle information in the form of words like 0000000 or 11111111 (or words from 0000000000000000 111111111111111111

Conventionally parallel buses are organised in multiples of four lines. and in microcomputing the most common bus-width is eight lines.

Binary display is easily achieved data on the bus is strobed into a latch and the contents of this latch are used to set up a display on eight LEDs. This project provides a small board with two displays to read an eight-bit word. If one display is all that is needed (to read a four-bit word) half of the design can be used just saw the PCB in half.

The data is loaded into a latch on the 9368 IC when the strobe line is taken low. This IC also contains the display drivers and all the electronics for decoding. The inputs are standard TTL-level and positive-logic ('0' < 0.4V, '1' > 2.4V).

The power-supply requirement is a single +5V.



Component overlay for the ETI 630 board shown below.

#### SPECIFICATION ETI 630

No of digits

Two

Number system

Hexadecimal (base 16).

Display format

0, 1, 2, 3, 4, 5, 6, 7, 8, 9, A, b, C, d, E, F.

Data input level

TTL positive logic.

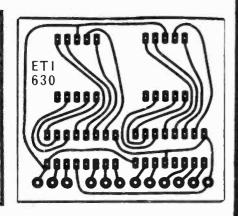
Strobe input level

TTL active low.

5 V. ± 0.25 V.

Power supply

Current consumption depends on display.





LAST MONTH I mentioned the new GI TV games chip which allows you to play up to six games on your TV by adding a few simple controls and components. This chip will only play "Ball and Paddle" games such as tennis, hockey and pelota, although the two Rifle games included on the chip allow shooting with a photo-sensor at the target (otherwise known as the ball).

GI have now released further details of their second generation of TV games including volleyball, tank warfare and a road race. Essentially the latest additions fall into two categories: multigame chips — which mate with other chips to increase the variety of games and cartridge programmable chip sets which use any or all of GIs lines of 16.8 and 4 bit MPUs.

All of the dedicated circuits are designed round a few basic chips, and with these you can build home TV games similar to those which are now becoming popular in arcades. In upgrading the standard "Ball and Paddle" games GI have added some interesting but basically useless gimmicks. These include differently coloured scores for each player, dual axis paddle control, and for squash, a ball which changes

With the new multigame chips the user can choose from four chips, one with the six basic paddle games, one with the simple games plus colour options, one with eight games or one with the tank battle alone.

colour to indicate whose turn it is to

With the add-on circuits you can add three variations of volleyball, a road race, a chase game or a battle between surface ships and submarines.

One of the other chips in this range is a five game chip which plays a set of games in which players have 16 directions to move

a vehicle or fire a missile, to make things as simple as possible GI have used simple squares to denote a player's pieces rather than using a tank shape or car shape.

For the higher priced game GI have a selection of games which use one of their MPUs to do a lot of the work, some of these games include blackjack, a slot machine, Noughts and Crosses and a Lunar Landing game. For further info you can write to GI at 63 Mortimer St, London W1.

#### **LANDED IN LUNAR-CY**

A game such as Lunar Lander can be played in three ways, as a picture game with a small rocket controlled by a joystick, as a numbers game with pictures given for height, speed, fuel, etc, or as a combination of the two. The "pretty pictures" version need a graphics generator whereas the alphanumeric only version relies wholly on your ability to read and understand the digital instrument readings given on a VDU. This second approach to TV games is more interesting for the home user as more brainpower is needed as opposed to the paddle game where skill and manual dexterity are of prime importance.

The object of a lunar landing game is to judge your landing speed and the amount of decelleration required at any given height. Decellerating uses fuel and only counteracts the effects of gravity. which causes acceleration, if you use too much fuel too soon you have none left for final manouvering and you land with a bump. If you do not use enough fuel early on then you cannot reduce your speed enough before impact, burn too much fuel when you are going slowly and you can end up going upwards. Add a few unseen calamities randomly such as fuel leak

suddenly losing you half of your supply and you have the makings of a rather good game.

#### **COUNTER MOVES**

The basis of the game is a set of digital counters giving the results from a simple algorithmic calculation based on changing input parameters. Sound complex? If you assume that your rocket passes into manual control at 5,000 feet up you have 100 units of fuel left and your present rate of fall is 150 feet per second it is relatively simple to work out the situation one second later.

If you do not fire your retrorockets then gravity will increase your speed by about 5 feet second to 155 ft/sec, also you have travelled for one second at 150 ft/sec (ignore increase in speed at present) and thus your new height is 4,850 feet. With no fuel burn, after ten seconds you will be at 3,275 feet and going down at 200 ft/sec, carry on at that rate and there won't be much left of you or your rocket.

#### **BURNING BOATS**

Let us assume that your retrorockets give you 10 ft/sec speed change in one second, thus if you now burn your fuel your speed is going to reduce by 10 ft/sec to 190 ft/sec and then to 180, 170, etc. Thus after another ten seconds your status is height = 1,725 feet, speed = 100 ft/sec and fuel = 90 units, the next ten seconds reduce these to 80 units of fuel at 1,175 feet with a speed of zero, if you continue to burn your fuel you will start going up and so you had better turn off the rockets.

After 15 seconds of free fall and five seconds of burn you will be down to 375 feet and going at 25

Seconds         Height         Speed         Fuel           0         5000         150         100           10         3275         200         100           20         1725         100         90           30         1175         0         80           40         950         50         80           50         375         25         75           51         350         30         75           52         320         35         75           53         285         25         74           54         260         30         74           55         230         35         74           56         195         25         73           57         170         30         73           58         140         35         73           59         105         25         72           60         80         15         71           61         65         20         71           62         45         10         70           63         35         15         70           64

hit it!

ft/sec. The table shows the final approach from this point onwards.

A landing at 10 ft/sec is like jumping off a ten foot wall, a lot better than landing at several hundreds of feet per second. By changing the parameters slightly and adding random fuel losses the game can become very exciting. The game is based on an algorithm (calculation) based on the changes in speed and fuel usage.

If you started out with only 30 units of fuel then the game would be more of a contest in fuel management, in the above example the 30 units would have run out after 63 seconds and you would have landed two seconds later but at 20 ft/sec.

#### **GAME AT MANAGING**

Other types of simple or complex management games of this type can be played with a MPU simply by programming it to do the calculations which can thus become more complex and have more random factors. For those of you who find the idea interesting there is a book by Hewlett-Packard called "What you do after you hit Return" (part no HP 36000-91005) obtainable from Hewlett-Packard Ltd.,

King St. Lane, Widdersh, Wokingham, Berks at £4.45 (Quote part number).

This book details the logic and a "BASIC" listing of about 50 games ranging from simple word and number games up to "Star Trader," Civil War and Stockmarket games. Comparing these games to the "Ball and Paddle" games is like comparing a spinning top to Monopoly, one is a short bit of fun suitable for pubs and arcades and the other is a multi-person competitive thinking game. Both types will be coming onto the consumer market later this year so start saving your pennies now!

#### **DON'T SWEAR TO IT!**

Some months ago we mentioned the problem of seven-segment to BCD conversion and several people came up with 3 or 4 chip solutions and even the idea of using a 74188 PROM, NatSemi have now announced (re-announced?) the DM86L25, "The one commercially available device specifically designed to do the job." It has two output enables, the ability to accept either positive or negative logic and the ability to recognise four letters, a minus sign and a blank — I wonder what the four letters are?

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