PE Tests New “Sonic Hologram” Preamp
Open Refrigerator-Door Alarm Saves Energy
Build an R-F Impedance Bridge

The Upcoming New World of TV Reception

Tested In This Issue
Eumig Model CCD Stereo Cassette Deck
Pioneer TVX-9500 TV Audio Tuner
Sabtronics 8100 Frequency Counter
The Fisher CR5150 cassette deck. Gorgeous up close. Even better from a distance.

Great styling and state-of-the-art performance are two things this new Fisher cassette deck has plenty of. But it's got something even more exciting: full-function remote control — without wires! It's the first tape deck ever to offer this sensational feature.

Think of it: by touching a button on the remote infrared transmitter, you can control Play, Record, Pause, Stop, Fast Forward, and Rewind modes — from up to 20 feet away! You can record, edit, search, and listen to cassettes — without leaving your easy chair.

And the CR5150 is just plain fun to operate.

Wireless control would probably make the CR5150 a big seller even if its performance was only average. But Fisher went all out, and gave it 3 heads for 30-19,000 Hz response, dual-process Dolby* for 68dB S/N ratio, and a servo-controlled transport with 0.04% wow & flutter (WRMS). Superb specs that only a handful of ultra-high priced cassette decks can match.

Feature-wise, there's a built-in digital clock that will turn on the CR5150 deck (or your receiver) to record anything you want at a preset time, whether you're home or not. The clock display doubles as an electronic tape counter with memory rewind. Silky-smooth, feather-touch buttons control the solenoid tape mechanism.

But considering the prices of other decks with similar performance and far less features, the Fisher CR5150 at $650* has to be one of the greatest values in high fidelity today. No matter how you look at it. Available at better audio stores or the audio departments of fine department stores.

"Manufacture's suggested retail value. Actual selling price determined solely by the individual Fisher dealer. New guide to buying high fidelity equipment. Send $2 for Fisher Handbook, with name and address to Fisher Corp., Dept. H, 21314 Lassen St., Chatsworth, CA 91311.

The first name in high fidelity.
Would you have bought?

The Digital Watch Hoax

Hundreds of consumers took part in an experiment. What it proved can be a lesson to us all.

The story we are about to tell you is the absolute truth. The persons involved, however, will not be mentioned in order to protect their reputations.

It began about six months ago when an enterprising watch manufacturer in Hong Kong started producing watches that were exact copies of the Seiko chronograph alarm selling in the United States for $300.

The Hong Kong version was sold to several American watch manufacturers for approximately $25. These companies in turn contacted several American mail order companies and offered the watch to them for around $50.

Soon, all you saw in newspapers and magazines were watches that looked like the Seiko but were selling for between $60 and $100. Although each watch had a different name, they were all made by the same manufacturer. Even JS&A was selling them.

Many of the other mail order companies had just started in business and were not financially stable. It service would ever be required and the company vanished, the watch would be useless.

A friend of ours, who was also in the mail order business, told us that for a low enough price the American consumer would be willing to buy anything, regardless of the stability of the company.

To prove his point, he made us an offer. If we could supply him with those digital watches, he would prove that the American consumer did not pay attention to who was offering the watch and only cared about price. We accepted his challenge. Prices had been dropping and the cost of our stainless steel model was now $38.

Our friend ran a small ad in the south-western edition of a consumer publication offering our watch for $39.95. The ad cost him only $72. It had no trial period, no accuracy claims and the name of the company used in the advertisement had never appeared before. His customers had to send in a check with their order and there was a $3.00 postage and handling charge. Even the name of the watch was not shown.

When the response came in, our friend was amazed. There were 38 orders, and he made a small profit. He delivered the watches and proved his point.

A smart consumer, however, would have never taken this gamble—at any price. He would have made sure the company was substantial, able to back their claims, and assure himself that the company would be around for awhile to service his purchase.

This ad was a hoax designed to prove the gullibility of the American consumer.

Indeed, not all those who responded were gullible. There were 62 letters from people who did not buy but asked for either more information, the name of the manufacturer, or the right to return the watch if it wasn’t any good. There were several inquiries made directly to the publication and a few to the Better Business Bureau.

There were more consumers who investigated the offer than those who took the bait despite the tempting price. That was encouraging for us.

But this story took an interesting twist after the experiment. The American watch companies handling the Hong Kong watch were getting overstocked. Prices were starting to drop and a few of the companies were indeed going out of business.

About this time, Texas Instruments introduced their new Chrono Alarm. It looked almost identical to the Seiko, but its features made it far superior.

The TI watch glows in the dark. A small tritium phosphor capsule, sealed by a laser beam, is located under the display. When the lights dim, the display appears to glow. You avoid the button pushing and component failures that are possible with watches that have miniature light bulbs inside.

And the features of the TI are the same as those of the Seiko plus a few more. The TI watch has a full-function chronograph, 12 or 24-hour time selectability, quartz accuracy to within 15 seconds per month, and a really fine quality case.

We felt that the TI Chrono Alarm had better quality and more features than anything else on the market, but it was priced at $125—higher than the Hong Kong watches.

So we tried another experiment. We offered the TI watch in a small advertisement in our catalog opposite an ad we created for the Hong Kong watch selling for $69.95. The TI watch generated four times the number of orders than the Hong Kong version. It was this test that convinced us to offer the TI watch in a national advertising campaign.

We are not showing you the Texas Instruments watch in this ad. First, it looks identical to the Seiko version and secondly, if we showed it and you just read the headline of this ad, you might think that the TI watch was “The Digital Watch Hoax” which of course it isn’t.

The TI watch will be sold in a few select stores shortly. Or you can order now directly from JS&A. We promise you prompt delivery and something even the stores don’t offer—the opportunity to wear the TI watch and the right to return it anytime within 30 days for a prompt and courteous refund if you are not absolutely satisfied.

If you are looking for the very finest watch you can buy—even better than the Seiko and backed by two substantial companies, we urge you to consider the TI Chrono Alarm. JS&A is America’s largest single source of space-age products—further assurance that your modest investment is well protected.

Send your check or money order for $125 for the stainless steel model or $150 for the gold-tone version plus $2.50 postage and handling (Illinois residents, please add 5% sales tax) to the address shown below. Credit card buyers may call our toll-free number below.

We will promptly ship your watch, one-year limited warranty and complete instructions. Then prove for yourself how outstanding the Texas Instruments Chrono Alarm really is.

There’s no gamble when you can own the finest. Order a Texas Instruments Chrono Alarm with complete confidence, at no obligation, today.

JS&A PRODUCTS THAT THINK

Dept PE One JS&A Plaza
Northbrook, Ill 60062 (312) 564-7000
Call TOLL-FREE 800 323-6400
In Illinois Call (312) 564-7000
©JS&A Group, Inc., 1979
PET 2040 Floppy Disk
Available May, 1979
List price $105.00/CE price $97.90

The Dual Drive PET Disk system is a second generation, with extremely large storage capacity and excellent file management system. Since this disk is an "intelligent" peripheral, it uses none of the ROM (read only memory) of the PET. The Floppy Disk operating system used with the PET computer enables a program to download data or write data in the background without simultaneously transferring data over the IEEE to the PET. The Floppy Disk system is designed for high speed data transfer. Due to the latest technological advances incorporated in this disk, a total of 3600 bytes are available in the two standard S-100 disks, without the problems of double tracking or double density. This is achieved by the use of two microprocessors built into the disk unit. Only two connections are necessary--an A.C. cord and PET interface. The 2040 Dual Drive Mini Floppy Disk Specifications
Mircrocomputer system devices

Now! 32K

PET 2001-32N Computer
Available May, 1979
List price $1,195.00/CE price $1,049.00

PET with 32K bytes of memory but has standard typewriter keyboards and no graphic keys. External cassette optional.

PET 2001-32 Computer
Available May, 1979
List price $1,195.00/CE price $1,049.00

PET with 32K bytes of memory and large key board with separate numeric pad. Graphics are on keys. External cassette optional.

PET 2001-16N Computer
Available May, 1979
List price $955.00/CE price $869.00

PET with 16K bytes of memory and large keyboard with separate numeric pad. Graphics are on keys. External cassette optional.

PET 2001-16B Computer
Available May, 1979
List price $955.00/CE price $869.00

PET with 16K bytes of memory and standard typewriter keyboards. External cassette optional.

PET 2001-8 Computer
Available May, 1979
List price $755.00/CE price $695.00

Our best selling personal computer. This is the original PET with integral cassette and calculator type keyboard. 8K bytes of memory.

PET 2001-32/16 SPECIFICATIONS
Dimensions: 16" wide by 16" deep, 14" high
Weight: 49 pounds (UPS shipping cost)

MEMORY
Random Access Memory 32K or 16K bytes
Read Only Memory operating system resident in the computer 14K bytes

CONTROL
100 character display monitor
VIDEO DISPLAY UNIT
9" monochrome CRT
8 x dot matrix characters
4K -- Operating system
16K -- BASIC monitor

KEYBOARD: Business or Number Pad Version
Screen Characters are top, bottom, left and right. Editing: Character insertion and deletion.
OPERATING SYSTEM: will support multiple languages.

Machine language accessibility. File management in operating system.

PET External Cassette
Available NOW!
at price $35.00/CE price $31.95
Cassette player/recorder to use with PET models 2001/32/16/5.

PET 2001-8 Computer
Available May, 1979
List price $755.00/CE price $695.00

PET 2001-16 Computer
Available May, 1979
List price $955.00/CE price $869.00

PET 2001-32 Computer
Available May, 1979
List price $1,195.00/CE price $1,049.00

PET 2001-32B Computer
Available May, 1979
List price $1,195.00/CE price $1,049.00

PET 2001-16N Computer
Available May, 1979
List price $955.00/CE price $869.00

PET 2001-16B Computer
Available May, 1979
List price $955.00/CE price $869.00

PET 2001-8 Computer
Available May, 1979
List price $755.00/CE price $695.00

Buy with confidence
To get the fastest delivery of your PET computer, send or call in your order directly to the Commodore Products Division. Be sure to calculate your price using the CE prices in this catalog. Most foreign orders are invited at slightly higher cost. International customers please request special shipping information (in our catalog) before you order. Contact your nearest retailer--they may have the latest computers in new condition with all accessories in 10 days, for a courteous and friendly explanation of the new computers, the PET is a very cost-effective unit with a built-in microprocessor. A retrofit kit is required for operation with the PET 2001-8 computer.

PET 2001 Single Drive Mini Floppy Disk
Available July, 1979
List price $559.00/CE price $539.00

The Mini Floppy Disk is a 80 dot matrix disk available from Commodore. The Mini Floppy Disk along with the PET 2001-32 or 2001-16 model is the basis of the new PET system. The Mini Floppy Disk has been designed for ease in operation by amateur and professional alike. The Mini Floppy is a complete unit to run the PET the first time with the best.
About the cover:

Tomorrow's standard TV receiver screen may be used for any of the diverse purposes shown on the cover—from light-pen drawings to facsimile printouts to extra channels for video newspapers, stock-market quotations, etc.

Cover Art by George Kelvin

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HOLOGRAPHY—VIDEO AND AUDIO

Some five years ago, I remember being impressed by a holographic image projected by a laser-beam system right here in our offices. It was a three-dimensional image of a woman in a glass jar, and I could observe a different view of the figure—left side, right side, front—as I moved around.

For whatever reason, visual holography has suddenly become a popular consumer subject in media and elsewhere. There’s even a company that’s marketing moving holographic images for promotional purposes (Man/Environment, Inc. Los Angeles, CA 90025). An article, “Holographic Memory,” in Psychology Today (February 1979) explores the principle of re-creation of an image in three-dimensional format from laser light patterns as it applies to distribution and storage in the human brain. The article focuses on visual aspects.

Three-dimensional audio has not been entirely ignored either. We had binaural recordings long ago, which, of course, require the use of headphones and special recording techniques that place microphones where ears would be located on a model of a human head. Four-channel sound was another effort to make sound reproduction more realistic through imparting a sense of depth. Bell Labs’ engineers, in turn, have experimented with pre-processing audio signals to speakers, using feedback loops to give a truer recreation of “concert hall” acoustics. Reports were that the effect was localized so that more than minor head movements would cause the enhanced stereo to be lost.

Other people have been working on audio systems that bear some resemblance to the Bell Labs’ experiments. One such system will soon be commercially available from Carver Electronics. A subjective report on the prototype model appears in this issue. The three-dimensional function is called a “sonic hologram.”

If not a true hologram, which consists of splitting a laser’s beam and reflecting part of it toward a film plate and another part to the object, with a reconstruction process taking place to create a holographic image, it comes close enough in an audio sense. Interestingly, I found the three-dimensional effect to be only moderately localized. This would seem to indicate either a more sophisticated system than the one Bell Labs used or that my neural holograms were peaked for the occasion.

According to the Psychology Today article, the brain has a capacity for a vast array of holograms. It observes that there are patch holograms, drawing an analogy with the hundreds of lenses of an insect’s eye that provide a composite image. It also points out that there are audio speaker systems composed of many small drivers that give the impression of a single image from one large driver. It further notes that an advantage to patches is that, when one moves across the control surface, the encoding is somewhat different, so movement can be sensed.

There’s obviously much more to the hologram subject than discovered at this time. But judging from the recent spate of “hologram” applications, scientists and others are gaining increased knowledge of the subject, all to our advantage. The “sonic hologram” is an example of this.
The Personal Computer Line by OHIO SCIENTIFIC

C1P: $349! A dramatic breakthrough in price and performance. Features OSI's ultra-fast BASIC-in-ROM, full graphics display capability, and large library of software on cassette and disk, including entertainment programs, personal finance, small business, and home applications. It's a complete programmable computer system ready to go. Just plug-in a video monitor or TV through an RF converter, and be up and running. 15K total memory including 8K BASIC and 4K RAM — expandable to 8K.

C1P MF: $995! First floppy disk based computer for under $1000! Same great features as the C1P plus more memory and instant program and data retrieval. Can be expanded to 32K static RAM and a second mini-floppy. It also supports a printer, modem, real time clock, and AC remote interface, as well as OS-65D V3.0 development disk operating system.

C2-4P: $598! The professional portable that has over 3-times the display capability of 1P's. Features 32 x 64 character display capability, graphics, full computer type keyboard, audio cassette port, and 4 slot BUS (only two used in base machine). It has 8K BASIC, 4K RAM, and can be expanded to 32K RAM, dual mini-floppies and a printer.

C2-4P MF: $1599! It's a big personal computing mini-floppy system at a special package price. Contains the famous C2-4P microcomputer with 20K static RAM, 5" mini-floppy unit for instant program and data loading, RS-232 circuitry (for optional modem and printer), and four diskettes featuring exciting games, personal, business and education applications.

C2-8P: $799! The personal class computer that can be expanded to a full business system. Has all the features of the C2-4P plus an 8 slot BUS (3-times greater expansion ability than the C2-4P). Can be expanded to 48K RAM, dual floppies, hard disk, printer and business software.

C2-8P DF: $2599! A full business system available at a personal computer price! The system includes the powerful C2-8P microcomputer (32K RAM expandable to 48K), dual 8" floppy unit (stores 8-times as much information as a mini-floppy), and 3 disks of personal, educational and small business applications software. Has all the capabilities of a personal system including graphics plus the ability to perform Accounting, Information Management, and Word Processing tasks for small business. Contact your local Ohio Scientific dealer.

Ohio Scientific offers a combination TV/Monitor (AC-3P) for $115.

* Monitors and cassette recorders not included. Ohio Scientific offers a combination TV/Monitor (AC-3P) for $115.

CIRCLE NO. 43 ON FREE INFORMATION CARD

Ohio Scientific
America's largest full-line microcomputer manufacturer
1333 S. CHILlicothe RD., AURORA, OHIO 44202 (216) 562-3101

AmericanRadioHistory.Com
Letters

FILLING IN THE BLANKS

I enjoyed the Electra "Bearcat 250" Scanner Product Test Report in the February 1979

issue but wish to note two errors. The first is that frequencies can be selected in 5-kHz steps only in the 146-to-174-MHz band. Programming increments are 10 kHz in the 32-to-50-MHz band and 25 kHz in the 420-to-470- and 470.0125-to-512.0125-MHz bands. Error number two is that an external relay is not necessarily required for controlling a tape recorder if the recorder uses a grounded-shield microphone or an auxiliary input plug. Incidentally, the scanner's i-f is at 10.85 MHz.--R.G. Borde, Sunnyvale, CA.

PHASE RESPONSE IS SIGNIFICANT

With reference to "Innovations in Speaker Design" (March 1979), at a propagation velocity of 1134 feet per second, a 1-ms delay is not 6° of phase shift for a 1000-Hz frequency as claimed. It is almost an entire cycle (317.5°, to be precise). It is very easy to rig a demonstration in which phase response does not matter. It is equally easy to devise a demonstration in which frequency response, distortion, or noise does not matter. Once critical listeners—mixers, musicians, etc.—become familiar with complete record/reproduce channels that are phase compensated from microphone to loudspeaker, they adamantly refuse to go back. Their ears have been trained to respond to a new dimension of fidelity. This is the best proof of the significance of phase response I can think of.

—Ted Uzzle, Cambridge, MA.

ASSIST WAS HELPFUL

Thanks for "Operation Assist." I received five replies to my published request. Respondents were most helpful. —Pastor Petersen, Langford, SD.

AM STEREO

We strongly object to "AM Stereo—Soon on the Air?" (December 1978) and particularly to the fact that the article was published without disclosing that author Joseph DeAngelo is an employee of the Broadcast Products Division of Harris Corporation, proprietor of the Harris CPM AM stereo system.

In light of the pending FCC proceeding concerning AM stereo broadcasting, we would expect Popular Electronics to be interested in presenting technically accurate and impartial information to its readers regarding all five proposed AM stereo systems. For example, although the article presents the Harris System as being superior, the FCC (Oct. 1978 release) expressed concern with certain limitations (stereo coverage area and separation) that appear to be inherent characteristics of the Harris System. Also, ABC has publicly indicated support for the Kahn/ Hazeltine AM Stereo System.—Edward A. Onders, Hazeltine Corp., Greenlawn, NY.

The author’s company affiliation was inadvertently omitted. Sorry.—Ed.

THE FUJI CHALLENGE

Try the others. Then try ours.

When it comes to choosing the best tape, a minute of listening will tell you more than hours of specs. Because the best tape for you depends solely on the sound you like and the response of your deck.

At Fuji, we make the most advanced magnetic tape in the world—for video as well as audio. We'll match our specs against anyone else's, but we respectfully suggest you stop reading and start listening. Once you compare Fuji FX-I or II to any other premium tape, there's nothing more to say. We have confidence in your ears.

FUJI

Magnetic Tape Division
of Fuji Photo Film U.S.A., Inc.
350 Fifth Avenue,
New York, New York 10001

Out of Tune

In the Audio Report on the Kenwood Model KT-917 FM Tuner (April 1979), in the subhead in the third column, a typographical error gave the input figure as "200,000 microvolts" when it should have been "200,000 microvolts."

In "The Morse-A-Word, Part One: Theory and System Operation" (March 1979), in Fig. 2, there should be a pin 9 on IC5 connected to pin 6 of IC11. The foil pattern for the PCB, shown as Fig. 6 in Part Two of the article in the April issue, supplies the necessary connection.

POPULAR ELECTRONICS
Burglar Alarm Breakthrough

A new computerized burglar alarm requires no installation and protects your home or business like a thousand dollar professional system.

It's a security system computer. You can now protect everything—windows, doors, walls, ceilings and floors with a near fail-safe system so advanced that it doesn't require installation.

The Midex 55 is a new motion-sensing computer. Switch it on and you place a harmless invisible energy beam through more than 5,000 cubic feet in your home. Whenever this beam detects motion, it sends a signal to the computer which interprets the cause of the motion and triggers an extremely loud alarm.

The system's alarm is so loud that it can cause pain—enough to drive an intruder out of your home before anything is stolen or destroyed and loud enough to alert neighbors to call the police.

The powerful optional blast horns can also be placed outside your home or office to warn your neighbors.

Unlike the complex and expensive commercial alarms that require sensors wired into every door or window, the Midex requires no sensors nor any other additional equipment other than your stereo speakers or an optional pair. Its beam actually penetrates walls to set up an electronic barrier against intrusion.

NO MORE FALSE ALARMS

The Midex is not triggered by noise, sound, temperature or humidity—just motion—and since a computer interprets the nature of the motion, the chances of a false alarm are very remote.

An experienced burglar can disarm an expensive security system or break into a home or office through a wall. Using a Midex system there is no way a burglar can penetrate the protection beam without triggering the loud alarm. Even if the burglar cuts off your power, the four-hour rechargeable battery pack will keep your unit triggered, ready to sense motion and sound an alarm.

ARRIVE HOME SAFE

There's personal danger in arriving home and finding a burglar in progress. And, if you surprise the burglar, you risk the chance of serious injury. With the Midex 55 protecting your home, you can open your front door with the confidence of knowing that no burglar lurks inside.

When the Midex senses an intruder, it remains silent for 20 seconds. It then sounds the alarm until the burglar leaves. One minute after the burglar leaves, the alarm shuts off and resets, once again ready to do its job. This shut-off feature, not found on many expensive systems, means that your alarm won't go wailing all night long while you're away. When your neighbors hear it, they'll know positively that there's trouble.

PROFESSIONAL SYSTEM

Midex is portable so it can be placed anywhere in your home. You simply connect it to your stereo speakers or attach the two optional blast horns.

Operating the Midex is as easy as its installation. To arm the unit, you remove a specially coded key. You now have 30 seconds to leave your premises. When you return, you enter and insert your key to disarm the unit. You have 20 seconds to do that. Each key is registered with Midex, and that number is kept in their vault that you should never need a duplicate. Three keys are supplied with each unit.

As an extra security measure, you can leave your unit on at night and place an optional panic button by your bed. But with all its optional features, the Midex system is complete, designed to protect you, your home and property just as it arrives in its well-protected carton.

The Midex 55 system is the latest electronic breakthrough by Sofian Systems, Inc. — a company that specializes in sophisticated professional security systems for banks and high security areas. JS&A first became acquainted with Midex after we were burglarized. At the time we owned an excellent security system, but the burglars went through a wait that could not have been protected by sensors. We then installed over $5,000 worth of the Midex commercial equipment in our warehouse. When Sofian Systems announced their intentions to market their units to consumers, we immediately offered our services.

COMPARED AGAINST OTHERS

In a recent issue of a leading consumer publication, there was a complete article written on the tests given security devices which were purchased in New York. The Midex 55 is not available in New York stores, but had it been compared, it would have been rated tops in space protection and protection against false alarms—two of the top criteria used to evaluate these systems. Don't be confused. There is no system under $1,000 that provides you with the same protection.

YOU JUDGE THE QUALITY

Will the Midex system ever fail? No product is perfect, but judge for yourself. All components used in the Midex system are of aerospace quality and of such high reliability that they pass the military standard 883 for thermal shock and burn-in. In short, they go through the same rugged tests and controls used on components in manned spacecrafts. Each component is first tested at extreme tolerances and then retested after assembly. The entire system is then put under full electrical loads at 150 degrees Fahrenheit for an entire week. If there is a defect, these tests will cause it to surface.

PEOPLE LIKE THE SYSTEM

Wally Schira, a scientist and former astronaut, says this about the Midex 55, "I know of no system that is as easy to use and provides such solid protection to the homeowner as the Midex. I would strongly recommend it to anyone. I am more than pleased with my unit."

Many more people can attest to the quality of this system, but the true test is how it performs in your home or office. That is why we provide a one month trial period. We give you the opportunity to see how fail-safe and easy to operate the Midex system is and how thoroughly it protects you and your loved ones.

Use the Midex for protection while you sleep and to protect your home while you're away or on vacation. Then after 30 days, if you're not convinced that the Midex is nearly fail-safe, easy to use, and can provide you with a security system that you can trust, return your unit and we'll be happy to send you a prompt and courteous refund. There is absolutely no obligation. JS&A has been serving the consumer for over a decade—further assurance that your investment is well protected.

To order your system, simply send your check in the amount of $199.95 (Illinois residents add 5% sales tax) to the address shown below. Credit card buyers may call our toll-free number below. There are no postage and handling charges. By return mail you will receive your system complete with all connections, easy to understand instructions and a one year limited warranty. If you do not have stereo speakers, you may order the optional blast horns at $39.95 each, and we recommend the purchase of two.

With the Midex 55, JS&A brings you: 1) A system built with such high quality that it complies with the same strict government standards used in the space program, 2) A system so advanced that it uses a computer to determine unauthorized entry, and 3) A way to buy the system, in complete confidence, without even being penalized for postage and handling charges if it's not exactly what you want. We couldn't provide you with a better opportunity to own a security system than right now.

Space-age technology has produced the ultimate personal security computer. Order your Midex 55 at no obligation, today.

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MAY 1979

AmericanRadioHistory.com
New Products

Additional information on new products covered in this section is available from the manufacturers. Either circle the item’s code number on the Free Information Card or write to the manufacturer at the address given.

Polk Audio
“Real-Time Array”
The “Real-Time Array Reference Monitor” Model R.T.A. 12 from Polk Audio claims, in addition to its frequency response of 27 to 20,500 Hz ±2 dB, the ability to pass recognizable square waves. Nominally rated at an impedance of 6 ohms, the system can be used with amplifiers in the range from 10 to 500 watts per channel and is said to produce a maximum sound pressure level of 120 dB with musical program material. The R.T.A. 12 uses a fourth-order electrical crossover between its two polymer laminate bass/midrange drivers and the one-inch soft-dome tweeter, and a fourth-order acoustic crossover between subwoofer and bass/midrange drivers. $350.

Circle No. 91 on Free Information Card

Alphanumeric Dot Matrix Printer
American Micro Products has introduced two new 5 x 8 dot matrix printers to provide alphanumeric hard-copy computer printout. Both printers include a general specification manual, parts list, flow chart, and schematics describing the 8-bit parallel interface (Centronics type). The 12-column PL12 printer is $59.95 and the 20-column PL20 is $99.95. Optional microprocessor control chip and printed circuit board for the interface are $99.95 and $29.95, respectively. (The board is available for the PL20 only.)

Circle No. 93 on Free Information Card

Antler Marine CB Antenna
Antler Antennas has introduced a new marine CB antenna to its line. The new “Sea-Sprite” antenna is a self-contained system that requires no ground plane. This allows the antenna to be used on fiberglass, wood, or metal boats, since the vessel is not part of the antenna system. The Sea-Sprite is a long-filament fiberglass stick. Its solid shaft is precision wound to provide efficient SWR/tri-gain characteristics and is sealed in a waterproof vinyl sheath. The molded base of the antenna hinges 180° to allow it to fold flat against the deck for out-of-the-way storage when not in use. For transmitting, the base snaps into a locked upright position. The antenna is provided with a factory-wired coaxial assembly. It is available in white, red, and black to accent or harmonize with boat colors. $28.95.

Circle No. 92 on Free Information Card

Audio-Technica Moving-Coil Phono Cartridge
Audio-Technica’s new Model AT30E moving-coil phono cartridge is said to overcome two traditional drawbacks commonly associated with such cartridges. First, it is relatively low in cost, retailing at a suggested price of only $100. Secondly, it is one of the few moving-coil cartridges to feature a user-replaceable stylus. High compliance is claimed to offer unusually high tracking ability combined with moderate tracking forces for longer record life. Offered optionally is the Model AT630 transformer ($95) that allows the cartridge to be matched to standard phono inputs.

Circle No. 94 on Free Information Card

Digital Wireless Security System
The Linear Corp.’s Model D-21 “Linear Alert” is a wireless security system in which a transmitter installed near a door or window and activated by any of a variety of sensor switches sends a coded digital radio signal to a receiver at a remote location. The receiver then activates an alarm.

(Continued on page 10)

OK Machine Digital Probe
OK Machine and Tool Corp.’s new PRB-1 digital logic probe is rated to detect pulses down to 10 ns and has a frequency range of 50 MHz. Automatic pulse stretching is to 50 ns (+ and −). It is compatible with RTL, DTL, HTL, TTL, MOS, CMOS and micro-

(Continued on page 10)
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MAY 1979

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CIRCLE NO. 22 ON FREE INFORMATION CARD

MAY 1979
NEW PRODUCTS (Continued from page 8)

processor logic families, and also features 120,000-ohm impedance, power-lead reversal protection, and overvoltage protection to ±70 volts dc. Its LEDs operate over a supply voltage range of 4 to 15 volts; an optional adapter can be used with supply voltages from 15 to 25 volts. Includes 6' coiled power cord and tip protector. $36.95.

CIRCLE NO. 95 ON FREE INFORMATION CARD

Apelco Marine Portable Radio

Apelco Marine Electronics’ Model AF-6 is a six-channel, hand-held, battery-operated vhf radiotelephone designed for use aboard small craft. It operates on eight penlight batteries or an optional nickel-cadmium power pack. Features include an internal telescoping antenna, an external antenna jack, and a meter for measuring transmitting voltage and signal strength. Comes with crystals for channels 6 and 16. Dimensions are 8½" x 3¼" x 1¼" (22 x 9 x 4.5 cm) and weight is 2 lb. $299. Address: Apelco Marine Electronics, 676 Island Pond Rd., Manchester, NH 01303.

CIRCLE NO. 96 ON FREE INFORMATION CARD

Storm Scan Alarm System

The Storm Scan Alarm System (Cat. #72,456) from Edmund Scientific is said to detect the approach of severe thunderstorms and tornadoes associated with high electrical activity as far as 30 miles away. A “lightning lamp” indicates electrical discharges by emitting green flashes, while a red burst lamp indicates intense electrical activity. It’s claimed that advance warning of 20 minutes is given on an approaching storm. A meter also indicates storm intensity and triggers audible alarm buzzer when the storm is near. The system measures 3" x 5" x 7" (8 x 13 x 18 cm), has a collapsible 18-inch antenna and weighs 3 lb. It operates from a 6-V battery. $89.95.

CIRCLE NO. 97 ON FREE INFORMATION CARD

Fujitsu Ten Car Speaker

The Model SSB-8G11 air-suspension speaker from Fujitsu Ten Corp. is designed for door installation. It has a detachable rain guard to shield it from internal moisture due to window leakage. The 4-inch, 8-ohm speaker is 1½ in. deep and each one is said to be capable of handling 20 Wrms. Features a cloth-rolled edge and wire-mesh grille. $42.95 per pair.

CIRCLE NO. 98 ON FREE INFORMATION CARD

Hand-Held 2-Meter Transceiver

Yaesu’s Model FT-202R is a hand-held 2-meter transceiver that provides up to six channels (three supplied and three optional) for two-way communication. Frequency coverage is 144 to 148 MHz, and r-f output power is rated at 1 watt. The double-conversion receiver section’s ratings include: 0.32μV sensitivity for 20 dB of quieting; −60 dB or better spurious radiation; ±20 kHz selectivity at −60 dB; 500 mW at 10% THD audio output. Features include: built-in speaker and condenser microphone; VOLUME, SQUELCH, and channel-selector controls; signal-strength/battery-condition meter. Power is from eight AA-size Ni-Cd or seven AA-size carbon-zinc cells (not supplied). Includes “Rubber Ducky” antenna and carrying case. Size is 17.1 X 67 X 49 cm, and weight is 400 g (about 14 oz). Address: Yaesu Electronics Corp., P.O. Box 498, Paramount, CA 90723.

Technics SU-C01 Preamp

Despite its minuscule dimensions of 11-11/16 by 1-15/16 by 9/4 inches, the Technics SU-C01 control preamplifier boasts a solid complement of control features and a head amp to accommodate moving-coil phono cartridges. Rated distortion for a 3-volt output is 0.005% worst case, and phono signal-to-noise ratios are 88 dB re 2.5 millivolts across 47,000 ohms for the moving-magnet section and 70 dB re 100 microvolts across 47 ohms for the moving-coil section. Phono equalization is said to be accurate within ±0.2 dB. High-level S/N is rated at 100 dB. Similarly sized companion pieces are the SE-C01 power amplifier and the ST-C01 AM/FM stereo tuner. $260.

CIRCLE NO. 99 ON FREE INFORMATION CARD

Digital Instruments with 3¼-Inch Readout Line

Pope Scientific has introduced a line of three “Digi-King” instruments which feature 3¼-inch digital readouts for reading at a distance. The first is actually a meter which will display any parameter capable of being expressed in millivolts from −1999 to +1999. The second is a thermometer which has a temperature sensor that can be used for surface, ambient, or immersion measurements and will display temperatures in Celsius, Kelvin, Fahrenheit or Rankine scales. It has an effective range of −55°C to +150°C (67°F to 302°F). The third instrument is a counter/ (Continued on page 12)
Integrated circuits are very private devices. When something goes wrong, they just don’t work. Which is tough enough when part or all of one IC goes bad. But often worse, because a single bad IC usually means a large, complex system that won’t function properly.

Until now, you could spend a lot of money and time—and still only be guessing what was happening at any point in a logic system.

Logic Probe LP-1. Captures pulses as fast as 50 nanoseconds, to 10MHz. Latching memory. Bargain-priced at only $44.95.*

CSC puts troubleshooting at your fingertips. Now, there’s a quicker, sure, less expensive way to get the information you need. CSC multi-family Logic Probes. Their LEDs light to show you at a glance the logic state at any point—and more. Catch fast pulses, even store them if you like. A flashing light signals pulse trains. And you can even approximate the duty cycle of asymmetrical waveforms.

Nothing could be simpler. No complex settings, no sync, no wait. A switch selects the proper logic family. The probes derive their power from the circuit under test. High input impedance prevents circuit loading. And all you do is touch the tip to any pin, pad or path for an instant picture of circuit conditions.

Laboratory quality. Economy price. High speed. High precision. Even memory. CSC Logic Probes deliver all the performance you need for design, development, debugging and servicing. Making digital work less of a chore, more of a bargain. CSC for yourself!

Logic Probe LP-2. All the basic features of LP-1, with pulses as fast as 300 nanoseconds, to 1.5MHz. Doesn’t have LP-1’s memory feature—but features even lower price: $24.95**.

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MAY 1979
Pioneer Non-switching Amplifier

Pioneer's Model SA-9800 integrated amplifier is built around new "Super-Linear RET's" (ring emitter transistors) in a non-switching design to reduce distortion. The amplifier is rated to deliver 100 W/channel continuously into 4 or 8 ohms from 10 to 20,000 Hz with no more than 0.005% THD. Frequency response is rated at 5 to 200,000 Hz +0/-2 dB, and S/N is 110, 90, and 72 dB through the AUX, moving-magnet, and moving-coil inputs, respectively. The amplifier features dual power supplies, full dc coupling, and a built-in head amplifier for moving-coil cartridges. Also featured is a fluorescent display that indicates peak power used (at 8 ohms) per channel in bar-graph form and indicates the program source selected; cartridge resistive and capacitive load selection; subsonic and high filter switches; two-position turnover tone controls; attenuator-type volume control; and independent tape monitor and duplicate switches. $750.

AP IC Test Clip Puller

The Super Grip II is a refined version of AP Products' basic IC test clip/puller. The new design features narrower nose clearance, which is claimed to permit easier attachment to ICs on densely packed boards. ICs with as little as 0.04" between adjacent legs can be tested.) A new "duck bill" contour has been added to the contact tips to make secure contact on DIP pins. Offset pin rows are provided to attach test probes to the clip, and "button heads" on the ends of the pins prevent probes from slipping off. Super Grip II test clips are available in 8-, 14-, 16-, 16- (LSI), 18-, 20-, 22-, 24-, 28-, 36-, and 40-pin configurations.

Pair up for perfect reproduction

With Osawa's Dynamic Duos

Advanced electronics are just not enough. For the most realistic music reproduction you need a high performance tonearm/cartridge combination to ensure smooth, wide frequency response and highly accurate tracking. The kind of top-quality performance that the Osawa duos deliver. Start with the UltraCraft AC-300 MK-II tonearm that features adjustable oil damping. This eliminates resonances by fine tuning the arm to your cartridge. Then select the Osawa cartridge that's priced right for you. You can mount it directly on the tonearm without a headshell, thus reducing effective mass to the minimum. Choose from 6 Satin Moving Coil models, the only MCs available that feature user-replaceable stylus and that don't require a step-up transformer or pre-pream. Or choose one of the newly introduced high-compliance Moving Permalloy (MP) models by Osawa. Top-of-the-line 300 MP offers a unique carbon-fiber cantilever that is extremely lightweight yet extraordinarily strong.

To select the combination that's ideal for you, visit your Osawa dealer. Ask him for the Osawa "Consumer Guide to Phonos Cartridges," and then get a complete demonstration. Whichever combination you select, the name Osawa assures you the ultimate in clean and faithful music reproduction.

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Or, plug in your Intra-Connector (see box) the same way, and you have an extra set of male contacts at right angles. Instant line-by-line probeability—and an easy way to tap your system and daisy chain it into new areas.

Both Intra-Connectors and Intra-Switches come in 20, 26, 34, 40 and 50-contact models.


Faster and Easier is what we're all about.
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Discovered by the U.S. Department of Agriculture, perfected by Pestolite; it actually draws over 300 different insects away from where you stay!

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The government discovered it. A lure like a magnet. Irresistible to mosquitoes, flies, moths, gnats, wasps and beetles. Over 300 annoying flying insects in all!

This discovery, by the U.S. Dept. of Agriculture at its field laboratory in Gainesville, Fla., was as timely as it was extraordinary. Because we also learned about the dangers of D.D.T. just about then.

Naturally, this discovery was public property. But there were problems that remained to be solved. What the U.S.D.A. had proved, beyond all doubt, was the fact that light sensitive, phototropic, insects would always respond to a particular kind of ultraviolet light. More, that the light of the extended far beyond the supposed ability of these insects to see. At the very minimum, one light could control an area as large as 1/3rd of an acre—14,250 square feet!

Foolproof. Safe. Silent.

The light attracted. But it didn't kill the insects. This is where Pestolite stepped in and created a simple, totally foolproof, completely safe and silent way to get rid of every bug attracted to the light. Without chemicals, electricity or polluting the environment.

Patio Protector can't harm your children or your pets. It just kills bugs. And so effectively it's approved by the Environmental Protection Agency, recognized by the Food & Drug Administration as well as the USDA even for use where food is prepared, in hospitals and commercial kitchens.

Low, Low Cost

Pestolite's achievement is notable in other extremely important ways. The other companies that used this discovery came up with bug killers that are no more effective; in fact,

*This is the area officially accepted as effective by the State of California! It signifies the tested and proved minimum insect control you can achieve under virtually any circumstances.

The End Of Them

Captured, unable to escape or fly away, the mosquitoes, moths, gnats and other insects are caught in a down-draft (created by a small electric fan) and plunged into the water in the pan below where they're drowned. All you do is change the water about once a week, emptying the tray in the bushes where the birds eat all the bugs. It's a clean, simple, second-week procedure.

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You can have the Patio Protector not for $199.95 or $160.00, but for just $45.00, plus $3.95 shipping and handling.

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Credit card holders may call the toll free number above. Or you can send your check made out to Douglas Dunhill to the address below. (Illinois residents include 5% sales tax.)

Designed to stand up to all kinds of weather, to provide years of trouble-free service, Patio Protector is covered by an unconditional one-year warranty. The special UV bulb will be replaced free if it fails for any reason—except negligent damage—within 6 months.

Order your Patio Protector today. It is a bargain at the price, but the pleasure of bug-free, bite-free summers around the house is priceless.

MAY 1979
New from NRI!

25" color TV that tunes by computer, programs an entire evening's entertainment.

Just part of NRI's training in servicing TV, stereo systems, video tape and disc players, car and portable radios.
Only NRI home training prepares you so thoroughly for the next great leap forward in TV and audio...digital systems. Already, top-of-the-line TV's feature digital tuning, computer programming is appearing, and new digital audio recording equipment is about to go on the market.

NRI is the only home study school to give you the actual "hands-on" training you need to handle servicing problems on tomorrow's electronic equipment. Because only NRI includes this designed-for-learning, 25" diagonal color TV with electronic tuning, built-in digital clock, and computer programmer as part of your training. With this advanced feature, you can pre-program an entire evening's entertainment...even key lock it in to control children's viewing.

As you assemble it, you learn how digital tuning systems work, how to adjust and service them. You work with the same advanced features used in the new programmable TV's and video tape recorders. It's exclusive NRI training that keeps you up with the leading edge of technology.

Exclusive Designed-for-learning Concept

The color TV you build as part of NRI's Master Course looks, operates, and performs like the very finest commercial sets. But behind that pretty picture is a unique designed-for-learning chassis...the only such unit in the world. Rather than retrofit lessons to a hobby kit or an already-built commercial set, NRI instructor/engineers have designed this television so each step of construction is a learning experience.

As you build it, you perform meaningful experiments. You see what makes each circuit work, what it does, how it interacts with other circuits. You even introduce defects, troubleshoot and correct them as you would in actual practice. And you end up with a magnificent, big-picture TV with advanced features. One you can sell or use in your home.

Also Build Stereo, Test Instruments

That's just a start. You demonstrate basic principles and circuits on the unique NRI Discovery Lab® then apply them as you assemble a fine AM/FM stereo receiver, complete with speakers. You also get practical experience as you build your own test instruments, including a 5" triggered sweep oscilloscope, CMOS digital frequency counter, color bar generator, and transistorized volt-ohm meter. Use them for learning, use them for earning as a full- or part-time TV, audio, and video systems technician.

Complete, Effective Training Includes Video Systems

Using NRI's exclusive methods, you learn far more than TV servicing. You'll be prepared to work with stereo systems, car radios, record and tape players, transistor radios, short-wave receivers, PA systems, musical instrument amplifiers, electronic TV games, even video tape recorders and tape or disc video players. Your training covers just about every kind of electronic entertainment equipment available now or in the near future.

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“Overall amplifier performance rating: excellent. Sound quality: superb.” Len Feldman*

When Len Feldman tested the Sansui AU-717 for Radio-Electronics a year ago, he concentrated primarily on its traditional, steady-state performance measurements. Power output capability. Total harmonic distortion. RIAA phono equalization accuracy. Signal-to-noise ratio. Usual tests, though applied to an unusual amplifier. Here's some of what he said:

“One clear advantage of DC design is apparent. Even at the low 20Hz extreme, the amplifier delivers a full 92 watts—the same value obtained for mid-frequency power—compared with its 85-watt rating into 8 ohms...

"The equalization characteristic of the pre-amplifier was one of the most precise we have ever measured, with the deviation from the standard RIAA playback curve never exceeding more than 0.1dB... The 380-mV overload figure for phono is far greater than would ever be required using even the highest output magnetic cartridges available."

At the time, dynamic response measurements—such as slew rate, rise time, and Transient Inter-Modulation distortion (TIM)—were still in their infancy. Indeed, even now, engineers have not yet fully agreed on a standard method of measuring TIM, though its audible effects have been increasingly recognized. Mr. Feldman sensed this when he commented: "Sansui claims that this unit has reduced transient intermodulation distortion...and, indeed, the model AU-717 delivered sound as transparent and clean as any we have heard from an integrated amplifier..."

The fact is that while conventional amps are designed to reproduce sine-wave test signals—which have a smoothly-changing, endlessly repeating character—with negligible THD, they usually do so at the cost of increased TIM. The excessive negative feedback used to reduce steady-state distortion to the vanishing point can (and usually does) reduce the ability of the amplifier to respond fully to the dynamic, rapidly-changing, pulsive signals which are the music itself. Thus, you get the harsh, metallic sound of TIM.

That's why Sansui has not only led the way in DC amplifier design (circuits whose low-frequency response extends down to zero Hz), but has also concentrated on the high slew rate, fast rise time designs needed for the faithful reproduction of music, not just simple test signals. Slew rate is a high 60V/µSec: rise time a fast 1.4µSec. And the frequency response of the power amp of the AU-717 extends to a full 200,000Hz.

Visit your authorized Sansui dealer. You'll hear the difference Len Feldman heard, and you'll understand why the Sansui AU-717 is about the most popular integrated amplifier available today.

Sansui AU-717
DC integrated amplifier.
THERE WERE claims that it was the biggest winter Consumer Electronics Show ever. Certainly it kept the city of Las Vegas spinning, as purveyors of transportation, lodging, and food tried to keep up with the traffic; and it seemed there wasn't a place in town where you could go to get away from it.

Still, actual new product introductions were not as numerous as is customary at the June CES in Chicago. For example, I counted only about 20 new receivers actually seen for the very first time—a substantial quantity to be sure, but not up to Chicago standards. (This may reflect the seasonal nature of the receiver market; receivers sell well at Christmastime, and the June show anticipates Christmas. Purchases of separates are much less seasonal.) Tandberg actually announced discontinued production on several receivers, retaining only the two top-of-line models.

Moreover, many of the receiver introductions were unexceptional mid-line models. Prominent among the stand-outs was the Bose "Spatial Control Receiver," designed specifically to interface with the later series of the manufacturer's 901 loudspeakers. The receiver employs four power-amplifier sections. Each of these drives one of the angled rear-panel driver arrays of the 901 systems. By varying the equalization applied to the two pairs of amplifiers, the stereo image can be broadened or narrowed. For use with other loudspeakers, the variable equalization feature can be disabled and the two pairs of amplifiers bridged for conventional stereo.

Other receivers of note included five new models from Nikko, headed by the 100-watt-per-channel NR-1219, an $800 125-watt unit (Model SA-5901) from Optonica, and an extension of the G-series Sansui receivers down to price levels of $320 and $270. Toshiba's recently introduced SA-7150 is the company's flagship model, with built-in Dolby noise reduction and phase-locked-loop frequency-synthesized tuning for both FM and AM. Rotel created a small sensation with the $310 RV-555, a diminutive "mini-system" with two small speakers and a vertically oriented receiver, all three in matched cabinets of virtually identical size. And Synergistics, like Bose best known for its loudspeakers, presented two new receivers.

Progress on the Metal Muddle?

This was to be the show that revealed the first widespread commitments to the technology of metal-particled tape, and it probably fulfilled its promise, all things considered. By this time most of the major tape manufacturers have responded to questions on their capability and willingness to produce metal tape with a guarded affirmative. In Las Vegas, however, several of them followed the lead of 3M (which initiated the metal furor last year with its emergent Metalfine product) and stood up to be officially counted as active supporters. At a pre-show press conference, Fuji announced full-production readiness, and privileged eyes were even allowed to gaze upon the first metal-partite TDK tapes seen publicly in the U.S. 3M freely distributed samples of blank Metalfine cassettes to all of us who asked directly, so that we could go crazy trying to record them. And Nakamichi gave out prerecorded metal cassettes bearing its own brand name, so that we could go crazy trying to erase them.

Still, the metal muddle goes on. Most manufacturers of software and hardware would prefer not to take a final step into production until they see some rationalized standards. And standards, except for the de facto or "emerging consensus" variety, are not yet with us. "We're demonstrating the potential of the technology, not the final product," is what most of the hardware manufacturers showing metal-ready machines have been saying. But what, in terms of the show, have they been doing?

By 3M's count, 16 manufacturers, all but one of which are major market forces here, showed metal-ready cassette decks at last October's Japan Audio Fair. It's interesting to note that well under half of these concerns openly displayed their machines in Las Vegas. (At least one brought his machine but kept it under close wraps in the back room. Another, Marantz, quietly pointed out metal-tape compatibility in one of its machines at last June's CES, but has remained pretty quiet about it ever since.) On the other hand, Las Vegas picked up three new metal-ready exhibitors: Nakamichi (Models 581 and 582), Eumig (Model FL-1000), and B.I.C. (Model T-4, which is the deluxe edition of this company's two-speed cassette decks, able to operate at 1¼ or 3¾ ips. In addition, B.I.C. demonstrated an under-dash two-speed car cassette player, the C1, which can handily play any tape the T-4 can record.) Also, Teac announced a retrofit program for the C-1 deck that will prepare it for metal at a cost of $150. Contact your nearest Teac service facility for more information.

Add to the above the shown-in-Japan Aiwa AD-6700, JVC KD-A8, Lux K-12 and 5K50, Onkyo TA-2080, Sanyo RD5372 and RD5370, and the Technics RS-M95, and we have a substantial turnout. This is not to mention the Tandberg cassette and open-reel machines that were the first to announce readiness for metal-particulate tape last year. (I have heard no other announcements for metal-ready open-reel decks—a pity, because the mid- and low-frequency capabilities of metal-particulate tape would seem capable of improving the open-reel format substantially.) It would be rash to list projected prices for these recorders, even when available, because they are perforce still prototypes. Suffice it to say that they are virtually all three-head machines sporting all the deluxe features the industry can offer; even the least of them will therefore be expensive. The two Nakamichi models are particularly interesting, not only for their asymmetrically balanced transports (intended to prevent additive effects of resonant modes), but also for the little mechanism that unceremoniously shoves the cassette's pressure pad out of the way when the transports are engaged. In a characteristically bold move, Nakamichi has decided that cassette pressure pads do more harm than good when precision tape guidance is desired. It will be interesting to see how the rest of the industry reacts.

May the Power Be With You. If you are a trend follower, there is a chance
the power won't be with you, because the hottest thing in Japan today seems to be teeny-tiny amplifiers, tuners, cassette decks, and speaker systems ("micro" or "mini" components) that, in the power-amplifier category, run out of breath at not much above 40 watts per channel maximum. Many of these units are so "mini" that their front panels can be shown full-size in a magazine such as this. The rationales behind these products seem to be various: cosmetic appeal (particularly for the so-called women's market); scaling down to fit diminished living spaces; a dramatic demonstration of today's electronics miniaturization; and perhaps a sudden shrewd realization that super audiophiles have been happily buying and using power amplifiers with minuscule power outputs because they believe in them for other reasons.

Last year Mitsubishi led the way with some charmingly colorful and ingeniously crafted little gems of this genre; now Technics and Toshiba are hard on their heels, with many more companies to follow if these icebreakers can warm up the market. Note that these products are not direct outgrowths of the low-silhouette rack-mounting components widely seen of late. The low, wide configuration permits open assembly on a single master circuit board, which eases production. The mini components bear the approximate dimensions and proportions of a college dictionary, and many of them are rather crammed and complex inside. The preamplifiers and tuners are undoubtedly state-of-the-art, or very close to it; the power amplifiers are still small in every sense of the word, despite the occasional use of switching power supplies.

Elsewhere in amplifiers, the show offered engineering philosophies aplenty. U.S. Pioneer is firmly attached to its "nonswitching" amplifier configuration and brought three integrated-amplifier examples along: the SA-9800 (100 watts, $750), SA-8800 (80 watts, $550), and the SA-7800 (65 watts, $450). Kenwood espouses high speed, and its new trio of integrateds (KA-907, 150 watts; KA-801, 110 watts; and KA-701, 80 watts) have prices between $1,000 and $450. The latest Sansui integrated amplifier, the AU-X1, also emphasizes speed; and at an output of 160 watts per channel and a slew rate of 260 volts per micro-second, it edges out the best of the Kenwoods by 10 watts and 30 microseconds. It also costs $450 more. Incidentally, the technologies employed in all these designs are complex, deserving of more expanded coverage than space permits here. The same is true of these manufacturers' latest tuners.

Manufacturers too numerous to mention, using the principle of floating bias point, are operating their output stages in class A at low levels and class AB at higher ones. Representative transition points are 5 watts (Rotel RB-2000) to 17.5 watts (Monogram 3100, 3200, and 3300). The latest Sony power amplifier, Model TA-N86B, makes no automated class transition. You can switch it to operate in class A for a maximum 18 watts out, class AB for 80 watts out, or bridged mono for 200 watts. Crown is more concerned with well-behaved protective circuits than exotic output stages, and much of the design effort for the new Model SA2 went into circuitry that closely monitors the operational state of the output transistors, limiting output only when safety is a factor.

Audionics and Infinity have meanwhile been thinking hard about driver stages rather than output stages. The answer they've both fastened upon—Audionics in the BA150 seen in prototype form last year and Infinity in the Hybrid Class A just being introduced—is vacuum tubes. Infinity's output stage uses transistors, and claims 150 watts per channel from a class-A configuration. Audionics gets the same output from a floating-bias vacuum-tube stage that is regulated by digital logic circuitry. Should you wish to go with tubes all the way, you no longer need turn to Audio Research or the other specialized companies. Word around the Show had it that Marantz intends to revive its classic Model 9 power amplifier and Model 7 preamplifier in kit form! These reports came to me a little late to check out, but even if there's some leg-pulling involved, the projected prices—$1,000+—might just persuade Marantz to sell you the necessary makings on special order. There was also a rumor that Marantz will offer a kit vacuum-tube tuner. If this turns out to be a revival of the justly legendary Marantz 10B, you are likely to have your kit-building and r-f-alignment skills taxed to the utmost.

Finally we come to one of the show highlights: the potential reconciliation of the mini-component phenomenon and the high-power trend. Carver Corporation's M-400 "Magnetic Field Amplifier" is rated at better than 200 watts per channel with more than acceptably low levels of noise and distortion. When it is understood that this performance is got-ten out of a basically unadorned cube 6¼ inches per side, with no sign of a heat sink, and with a weight of about 12 pounds, it gives one pause.

After scrutiny of a rough block schematic and a conversation or two with the designer (who was secretive at the time because of pending patents), I propose that the magnetic-field amplifier works like this. Power enters from the wall socket, encounters heavy-duty devices that switch it to an appropriately high ultrasonic frequency, and then subjects it to push-pull amplitude modulation by the incoming audio signal. Then, and only then, does it run up against any sort of power transformer, which in this case is diminutive compared with those devices that must handle high currents at 60 Hz. The output of the transformer, stripped of its ultrasonic carrier, is perfectly capable of driving loudspeakers directly, although distortion would rise to undesirable levels at higher frequencies if it were done this way. This problem is solved in a manner yet unclear, and the amplifier, from all indications, proceeds to work. Intended asking price: approximately $300.

Among other innovations from Carver Corp. is the "Sonic Hologram Generator" preamplifier, which incorporates in one unit all the heralded innovations that its designer has brought to signal processing over the years, plus a "hologram generator," which processes a stereo signal (at least it has no effect on a mono signal) in a way that was empirically derived, according to Bob Carver. I am told that a report on a prototype of this product will appear elsewhere in this issue, so I'll refrain from comment here. But after an extended listening session with program material of my choice, I found much that sounded attractive and nothing wrong that I'd care to stake my typewriter on verifying.

The Last Gasp. As this column comes breathlessly to a halt, apologies must be made for all that has been left out. Record players at the show appeared to be getting cheaper and in many cases better; moving-coil phono cartridges continue to prosper and proliferate; tape machines of all formats are, or can be made, more and more suited to the tapes you can actually buy today; loudspeakers, as usual, remain unknown quantities until they can be auditioned at length in congenial surroundings; and what is happening to high fidelity in automobiles is hardy to be believed. But we will save this for later.
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A STEP BEYOND STEREO
A Preview of the Carver C4000 Preamp

By Bob Houck

T O CALL the Carver C-4000 merely a preamplifier would be to lose sight of the underlying philosophy of its design. Actually it is much more than an RIAA phono stage and power-amplifier controller. This component is so replete with signal-processing functions that the designer obviously considers operations of this kind to be highly desirable, if not actually essential.

In addition to the customary RIAA equalization required for disc playback and optional equalization via gain and tone controls, the Model C-4000 contains peak-unlimiter/expander circuitry to enhance dynamic range, an “autocorrelator” to reduce noise in the program material, three channels of time delay (two real and one derived from the others), and a startlingly new function that has been dubbed a “sonic hologram generator.” Also included are two power amplifiers for the delay channels, each capable of delivering about 25 watts to an 8-ohm load.

The C-4000 includes two phono inputs and provisions for two tape recorders with dubbing possible in both directions. Tone controls are of the usual bass and treble variety with treble turnover frequencies switchable between 8 kHz and 2 kHz and the bass operating with either a 40-Hz turnover or as a loudness control. Tone controls are defeatable and can be set separately for each channel. An additional EQ function available is a 2-dB shelving cut from 400 Hz downward. An infrasonic filter and 20 dB of output muting are also included, and if more signal processing is desired there is an external loop.

Of all the signal-processing modules included in the C-4000, the sonic hologram generator is the most exciting and innovative. Other than the determination by Hirsch-Houck Labs that the hologram added nothing significant in the way of harmonics (better than 84 dB below 0.5 volt output at 1 kHz) or noise, evaluation of the device was accomplished by listening tests. The purpose of this function is to broaden the stereo image, increase the precision of localization of individual sound sources, and extract ambience in a manner reminiscent of quadraphonics (except that one pair of speakers is used).

**Sonic Hologram Listening Tests.** The intensity of the hologram effect and the listening positions from which it can be perceived depend strongly on the characteristics of the listening room and the loudspeakers. But in any case, sound realism is enhanced. Because the hologram is new and fairly radical in concept, subjective evaluation of it presented a conundrum of considerable proportions. In a sense, we had to develop an understanding of what it was supposed to do before we could tell how well it actually performed. Bob Carver, the designer, was lucid enough about his intentions, but it soon became clear that not even he understands the hologram completely. Moreover, the effect of the device is such that its own description of itself, if one may resort to such hyperbole, is more eloquent than anything Carver could say. Our experience with the hologram was, then, as much a learning situation as an evaluation, for what happened at each stage suggested new paths to be explored.

We heard the hologram first in a moderately large room (room A) with the listening posi-
tion roughly 10 feet from a pair of front-firing loudspeakers designed for minimum time dispersion, and standing about seven feet apart in the center of the room. We were immediately struck by the effect. The music took on a pleasing three-dimensional quality and the performers seemed to be enveloped in a real space as long as the listener was located fairly precisely on the center line between the speakers. Movement of one’s head more than a few inches from side to side weakened the illusion considerably; movement of a foot or two reduced it virtually to ordinary stereo.

A second test took place in a fairly small room (room B), also using front-firing speakers, but without phase correction. Here the results, though easily perceptible, were less striking, while the listening position was about as critical as before. Curiously, the hologram seemed to correct a deficiency of bass in the speakers. A second pair of speakers containing drivers angled toward the side was substituted, and the effect was further diluted to the point where it was almost disappointing. In this instance, however, the listening position, however, was less critical.

Yet another trial took place in a moderately sized rectangular room (room C), with the speakers standing rather far apart on the long wall. The speakers were somewhat unusual in that the woofer and tweeter were aimed forward while the midrange driver radiated to the rear. Once again the effect was modest in its intensity, though the tolerance for variation in the listening position became broad enough for two people seated side by side to perceive what little there was. Another odd side effect was that the music seemed louder when the hologram was switched in, which a sound-level meter appeared to confirm.

The final and, in our view, most successful test took place in a fairly large listening room (room D), using speakers without correction for time dispersion. Both speakers were set well away from the room boundaries, and a variety of listening positions were tried. In general, the hologram gave a pronounced and vivid impression at a listening position on the center line between the speakers and some 12 feet away. The effect was still most intense at the center position, but there was less deterioration with lateral movement.

Curiously, we noticed that stereo image broadened more to the right side than to the left, an anomaly that we finally traced to a wall reflection. To minimize it we resorted to a more “classical” geometry, with the listening position and those of the loudspeakers forming an equilateral triangle. Using this arrangement, the result was positively breathtaking! When the lights were turned out, we could almost have sworn that we were in the presence of a real, live orchestra. An additional bonus was that two listeners side by side could enjoy the effect.

In most of the tests the time-delay system, with the left and right auxiliary speakers positioned to the sides approximately in the plane of the listener’s ears and the center channel midway between the main pair and a little farther away, was synergistic with the hologram.

It enhanced the sense of depth and enlarged the perceived space somewhat. When the listener was close enough to the main speakers to minimize room reflections, however, the contribution of the delay was small.

Those who live in small apartments with thin walls will be interested in another benefit that the hologram provides. An obvious solution to the problem of listening without disturbing neighbors is to position minispeakers close to one’s ears and listen at a level that, though subjectively fairly loud, is moderate at any distance. Unfortunately, this usually produces some of the problems associated with headphone listening, such as sounds coming from inside the head. The hologram clears this up very nicely and makes for a thoroughly realistic listening experience.

How it works. Some insight into why the hologram acts as it does can be obtained from examination of its inner workings. Briefly, what the system does is to anticipate the amount of crosstalk reaching the left ear from the right channel and the right ear from the left channel and introduce corresponding cancellation signals. In essence, each channel receives an admixture of a phase-inverted and slightly delayed version of the contents of the other channel. The crosstalk, it should be made clear, is not a result of electrical deficiencies but of sound from the right speaker reaching the left ear and vice versa.

If the cancellation signals are mathematically correct, Carver says, the system works only if the listener’s head remains absolutely fixed on the center line between the speakers, a result similar to that produced by a "binaural listening system for loudspeakers" demonstrated in prototype by another company. What Carver has done is to trade away a small—and seemingly negligible—part of the effect to allow the listener to move his head.

"With the lights out, we could have sworn we were in the presence of a real-live orchestra."
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and under optimum conditions move slightly off center. Interestingly, Carver's optimization was achieved empirically rather than by calculation, but whatever its genesis, it works.

**Lab Tests.** Since the Carver C-4000 is still in preproduction stages, Hirsch-Houck Labs tested a prototype that, while it gave generally excellent results, will undergo additional refinements before units are offered for sale. The lab was able to verify that all inputs are adequately sensitive and quiet and that the overload margin of the phono stage is sufficient. Harmonic distortion measured quite low at output signals up to 0.3 volt, and spectacularly low with signals up to 6 volts. (The nominal maximum output is 2.5 volts.) Signal-to-noise ratio was 74 dB (IHF "A" weighted) re 0.5-volt output through a high-level input and 70 dB through the phono input re 2.5-millivolts input—very good performance in both cases.

Phono equalization is accurate to within ± 1 dB across the audio band. The tone controls behave essentially as they should, although, somewhat unusually, they allow approximately 4 dB more boost than cut at extremes of the spectrum. Response of the 400-Hz shelving circuit is satisfactory, though the level at that frequency is down just 1 dB; the full 2-dB reduction is achieved by 200 Hz and is unchanged down to 20 Hz. A check of the phono stages for cartridge interaction showed that this effect is not present in any significant degree with three popular cartridges.

Analysis of the time-delay section indicated some problems with alias distortion—essentially intermodulation with the switching frequency used to shift the signal along the delay line. Bob Carver explains that the delays used in the prototype are too long—some 50 or 85 milliseconds, at the user's option. This results in a switching frequency that is a little too low. The net result is the aliasing problem, which occurs despite the use of a low-pass filter with an extremely sharp rolloff at the delay line input. Besides that, the sharp cutoff of the filter causes ripple in the passband, another source of coloration in the sound. The solution, which Bob Carver says he will apply in production models, is to reduce the delay times and speed up the switching frequency by about 40%. This should reduce aliasing to insignificant levels and at the same time relax demands on the filter, allowing a gentler cutoff and smoother passband.

The power amps for the delay line outputs showed measured THD on the order of 1% at all signal levels checked. Considering that these are provided more or less as a low-power-output convenience, their performance is certainly adequate for the purpose. Users who demand more pristine delayed signals can use the direct outputs to drive power amps of their choice.

**The Other Processors.** Action of the peak unlimiter/expander section is carried out by three subsections. The first, which comes into play at a nominal 0-dB level is a fast-attack/fast-release circuit that increases gain for peaks by 1.5 dB. Over the range down to -10 dB the signal passes unchanged. Between -10 and -20 dB, the signal is expanded downward by 2.8 dB, using a circuit that acts slowly in reducing gain and quickly in increasing it. Below -20 dB a circuit that acts slowly in both directions gives another 2 dB of downward expansion. The two downward expanders interact (their resistive elements are connected in parallel to form the shunt leg of a voltage divider) so that their relative gains depend to a degree on which acts first. This property, coupled with the different attack and release times and the modest ratios of expansion, produce a system that rarely, if ever, betrays its operation. The current expander/unlimiter, says Carver, represents an improvement over his earlier designs.

Similarly upgraded, according to its designer, is the "autocorrelator" noise reduction circuit. Where earlier designs strove for 10 dB of noise reduction, the present one settles for 8 dB in order to be less audible in its operation.

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**Phono equalization is accurate to within ±1 dB across the audio band.**

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**Fig. 1. Schematic of control circuit for one of the transmission gates of the autocorrelator.**
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Fig. 2. Spectrum analysis of a 15-kHz sine wave mixed with pink noise. The upper trace is a reference level established with the autocorrelator switched off. With the autocorrelator on (lower trace), the noise is reduced by 8 dB except for the immediate neighborhood of the 15-kHz tone and at low frequencies.

One problem that plagued the earlier version was its compromise attack/release time. If the system acts rapidly enough to prevent audible "breathing" through modulation of the noise level as the signal decays, changes in the signal envelope—as in a vibrato or tremolo—are also capable of modulating the noise and producing a coloration that can seriously affect the sound of a solo flute or violin. Carver overcomes this with a control circuit that attacks differently for large and small signals.

In Fig. 1, which shows control circuitry for one of the six transmission gates used in the unit, audio, after having traversed an appropriate band-pass filter, is buffered by A1, passes through R3, and is rectified by D1. Capacitor C1 smoothes the resultant dc, which charges C2 through R2 and passes on through R4 to A2. This, in turn, feeds a control voltage to the gate that passes audio that is present at the input of A1 to the system output. Time constant R2C2 is long, so that the circuit operates slowly. When the dc level across R1 and C1 increases to a point where the barrier potential of D2 is overcome, the diode conducts and bypasses R2, allowing C2 to charge rapidly. Thus, the control voltage reacts slowly to small signals, rapidly to large ones.

When audio levels decrease, C1 discharges rapidly through R1 and C2 discharges slowly through R1 and R2 in series. If the difference between the voltage across C1 and C2 is enough to forward-bias of D3, R2 is again bypassed, which speeds up the discharge of C2. Once again, the circuit responds quickly to large signals, slowly to small signals. Diodes D4 through D7 clamp the control voltage so that the gate is not overbiased. One gate operates from 200 Hz.

Comments from the listening panel...

This component might well be called an "omnibus preamp" because it contains several signal processing stages not ordinarily found in a preamplifier. Of these, it is the "sonic hologram" circuit that is really the star of the show. The circuit processes only the front-channel signals. For maximum realism, however, it was my experience that the sonic hologram worked best when the time-delay system was driving three small speakers (one connected across the internal auxiliary amplifier's "hot" terminals). This is based on auditioning the equipment in a small room (room B) and in a medium-size room (C) with reflective speakers on the long wall.

Frankly, I tended to be skeptical about the ability of a device to substantially improve the spatiality of a sound field generated by only two speakers. But after listening to the Carver preamp connected to two different audio systems, I'm a believer! I found it simply amazing to hear an instrument apparently located several feet to the right and a few feet forward of a right-channel speaker. The sonic hologram circuit adds dimensions to sound systems that they have never had. Moreover, its effect is convincing, not artificial.

The principal limitation is that of the area of maximum effect. Both the degree of improvement and size of the area of maximum effect varied in the two listening environments in which I heard the sonic hologram operate. In one room (B), the area (and degree) of maximum effect was fairly small. In the other (C), the area was larger, as was the degree of improvement.

In the second room, a more noticeable increase in average sound level was discerned. In both, the frequency response of the speakers (different units in each case) seemed broader than before the circuit was switched in. A serious deficiency in bass response of the first room's speaker was apparently rectified—even though the sonic hologram contains no boost in itself.

Accordingly, it's impossible to predict exactly what will happen when the preamplifier is connected to your system. I can state with a high degree of certainty that your system will sound better than it ever has, but a lot of experimentation with speaker placement (and listening location) is warranted.

The rest of the preamp is fine, though not as dramatic as the sonic hologram unit. As a whole, it offers a high level of performance and the flexibility that those who purchase separate components seek. At first glance, the preamplifier's price tag might seem steep. However, this is much more than a preamp, and yields audible dividends that are more than justified the investment it requires.

—John J. McVeigh, Technical Editor

Carver's new preamp prototype is certainly a bold stroke of design innovation, combining a host of special signal-processing features, the most dramatic being its "sonic hologram." Listening to its performance for some 15 hours in three different locations, each with different equipment, convinces me that this new function indeed adds a sense of viv-
downward. Five gates divide the interval between 1.5 kHz and 20 kHz into equal fractions of an octave.

Capacitor C3 represents a bit of psycho-acoustic trickery played on the ear so that high frequencies do not appear to be lost when the control voltage is not sufficient to open the gate. Some of the ripple from the rectified audio passes through C3 and is fed to the gate along with the control signal. Resistor R4 provides isolation so that C2 does not short out the ripple. When the gate is closed, a very small amount of ripple (a highly distorted version of the audio) bypasses the gate and goes to the system output. The level is so low that the ripple cannot be identified as such, but it substitutes for high frequencies that might otherwise be missed. When the gate is open, this low-level contribution is masked. The optimum result is achieved when the audio level at the input of A1 is scaled so that positive-going audio can pass through D1 while positive-going noise cannot. A spectrum analysis of noise reduction in the autocorrelator is shown in Fig. 2.

Audibly, we were not able to detect the operation of the autocorrelator other than through the reduction in hiss. Flute and violin sounds in particular were very well handled. The system works best at reducing low-level noise yet further, but if its threshold is suitably adjusted, it will help noisy program material too. The automatic threshold adjustment provided works well for most normal material. It may be possible for some programs to "catch the system out." But none of the fairly wide variety of selections we used did.

From where we sit, the Carver C-4000 looks like one of the most interesting and provocative components to come along in some time. The peak unlimited/expander, autocorrelator, and time-delay selections (with built-in power amps) are useful adjuncts to the highly competent basic circuitry. As such, they are most welcome. But the real star of the show is the sonic hologram, which constitutes a significant step forward in stereo imagery and the recreation of a sense of space. Another pleasant surprise, considering the unit's multiplicity of features, is its projected price, which is in the $800 to $900 range.

—Harold A. Rodgers, Senior Editor

id realism to sound reproduction.

The degree to which it enhances sound depends on many factors, though. For example, in two locations (rooms C and D) the listening position between speakers was not especially critical as far as getting the full "hologram" effect. But in one situation (room A), more than minor head shifts caused a marked dropoff in the effect. Program material makes a difference, too, as does the type and positioning of speakers. And, interestingly, so did a switch of cartridges. A Denon moving-coil model coupled with a Mark Levinson pre-preamp, for instance, produced more sound spaciousness with ordinary stereo than did a Shure Type V-15 IV. But the converse was true when the hologram was activated. I would recommend, too, that a "minispeaker" be used for center channel purposes, as it adds noticeably to a three-dimensional effect in some circumstances.

Adding sound delay to the "hologram" did not contribute any sound change in many instances. I must confess, though, in ordinary stereo it clearly improved sound quality. The same held true, to a lesser degree, with the expander function. I found.

In sum, listening to the C-4000 in its hologram position was a thrilling sound experience for me. Plain old stereo will never be the same. I do hope, though, that Carver has plans down the pike to offer the "sonic hologram" generator in a less complete package, that is, the sonic hologram and sound delay functions only.

—Arthur Salsberg, Editorial Director

Sitting in the optimum location and listening at first in conventional stereo, I found that switching in the sonic hologram generator made a difference that it would be a gross understatement to call dramatic. The sound became rich and solid, and there was a great apparent increase in bass. Instrumental sounds that had been confined to the region between the speakers suddenly stretched along the sides of the room nearly as far back as I was sitting (some 12 feet from the speakers). Adding the delay channel extended the richness and depth all around me. When the delay and the hologram generator were switched out, the sound collapsed to ordinary stereo and was a pronounced letdown.

As I moved away from the optimum position, the almost palpable sense of reality first weakened and then disappeared. What did remain was the strengthened bass and a sense of ambiance that seemed at least as good as anything time-delay devices had produced in my listening room (B). Even though I knew only two speakers were in use, the fact was hard to accept.

On the basis of this brief exposure to the Carver C-4000, I am eagerly awaiting a production model that I can listen to at length. And I can only suggest that any audiophile who is skeptical of my glowing enthusiasm get one to hear for himself. The effect strains credibility—had I not experienced it, I probably would not believe it myself.

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John E. Cunningham
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The Technics isolated-loop system. It's the one big difference between their decks and ours.

Every one of Technics four open reel decks has one thing in common: The performance of Technics isolated-loop tape transport system. And that means performance that's comparable to professional open reel decks costing thousands of dollars more.

By isolating the tape from external influences, our isolated-loop tape transport system minimizes tape tension to a constant 80 grams. This not only provides extremely stable tape transport and low head wear, it also reduces modulation noise and wow and flutter to the point where they're detectable on only sophisticated testing equipment.

Electronically, our line of isolated-loop tape decks are equally impressive. The reasons are as simple as their full logic transport controls, highly accurate microphone amplifiers, FET mixing amplifiers and separate 3-position bias/EQ selectors.

And you'll get all this technology whether you choose the two-track RS-1500, the extended playing time of the 4-track RS-1700 with auto-reverse or the studio features of the RS-1520.

There's also an optional full-feature infrared wireless remote control (RP-07C; With it you can get your hands on all this sophistication from up to 20 feet.

All four decks hit the competition right between the reels. Because all four have: REG. RESP: 30-30,000 Hz, ± 3 dB (−10 dB rec. level) at 1.5 ips, WOW & FLUTTER: 0.018% WRMS at 1.5 ips, S/N RATIO: 57 dB (1506 & 1700) and 60 dB (1506 & 1520) NAB weighted at 1.5 ips. SEPARATION: Better than 50 dB. START-UP TIME: 0.7 sec. SPEED DEVIATION: ± 0.1% with 1.0 or 1.5 mil tape at 1.5 ips. SPEED FLUCTUATION: 0.05% with 1.0 or 1.5 mil tape at 1.5 ips. PITCH CONTROL: ± 6%


Technics Professional Series

CIRCLE NO.56 FOR FREE INFORMATION CARD.
Eumig 'CCD' Cassette Deck

The new Austrian-made Model CCD from Eumig is a three-head, top-loading cassette deck with many unusual design and operating features. In particular, it has a low-mass servo-controlled capstan motor with optical feedback, solenoid control of all transport functions, full remote control as a standard feature, flexible dual inputs with mixing capability, and fast peak-damped motion of level indicators.

The recorder is finished in black and has an upward-sloping rear section that contains the level indicators and ON/OFF power switch. The transport controls are light-touch pushbuttons, each of which has a status indicating LED near it. Input and output jacks are in the rear, microphone jacks are in the front, and various controls and switches are located on the front and underneath the recorder.

The Eumig Model CCD measures 17.1" W x 11.8" D x 5.4" H (43.4 x 30 x 13.8 cm) and weighs 16 lb (7.3 kg). A remote-control accessory supplied with the recorder duplicates all transport-control functions, including recording and fast forward and reverse. It measures about 3.5" x 2" x 1" thick (7.6 x 5.1 x 2.5 cm) and plugs into the recorder through an approximately 90" long (2.3 m) cable. Suggested price is $1300.

**General Description.** A light touch on a button opens the door of the cassette compartment, while an additional, firmer pressure causes the cassette tray to tilt upward with a slow, damped motion for loading and unloading a cassette. Selection of bias and recording and playback equalization is automatic, using the special keying hole on the rear of a cassette to select operating conditions for chrome (Cr) tape; otherwise, the machine is set up for either ferric-oxide (Fe) or ferrichrome (FeCr) tape. A slide switch under the cassette cover permits the user to choose between these two tapes, since there is no special cassette keying to distinguish between them. The appropriate LED in front of the cassette compartment comes on to indicate the type of tape being played or recorded.

The deck has a memory-stop feature associated with its index counter. The stop works in either direction of tape motion in the fast-speed modes when the counter reaches 000. Two pairs of slide-type potentiometers are provided for recording level adjustments on each channel for each of the two inputs. An input is selected by touching a button in front of its level controls. Simultaneously, the scale markings for the selected controls light up and the appropriate INPUT LED lights up. Normally, touching the other input button will shut off the first input. However, both inputs can be used simultaneously by holding down one button while touching the other.

Between the INPUT select buttons is another button labelled AUTO/MAN. Touching this button causes the manual level controls to be replaced with an automatic volume control (avc) circuit. This function would principally be used for speech recording. It can be used with either recording input but not with both simultaneously. When the avc is on, the lights for the level controls and the LED indicators extinguish. Normal manual control is restored with a second touch of the AUTO/MAN button.

Plugging a microphone into one of the input jacks replaces the corresponding LINE input of INPUT 2 with the microphone signal.

The recording-level indicators are calibrated from -20 to +6 dB with LEDs that form a line whose length is proportional to the peak incoming signal level. Up to -3 dB, the LEDs are green; the 0-dB LED is yellow; and the +3- and +6-dB LEDs are red. Since the signal level is monitored after the recording equalization, the LEDs give a true picture of the peak levels applied to the recording head.

The transport controls are fully logic operated through solenoids. Any transport control button can be touched while the machine is running in any mode without risking damage to tape or deck. Even the button for the cassette compartment door can be operated while the tape is running. Pressing the OPEN button while the deck is operating causes the tape to stop. Then the compartment door opens and allows the tape to be lifted out of the deck by the tray. Only the record function cannot be engaged once the tape is in motion. The deck is placed in the record mode in a rather unorthodox manner. The PAUSE button must be touched first, then the REC button. After the levels have been set, another touch of the PAUSE button starts the recording process. Alternatively, the REC button can be pressed first, followed by two touches on the PAUSE button. While
recording, touching PLAY will instant-
ly place the deck in the play mode.

As befits a true three-head record-
er, the Eumig CCD has full off-the-
tape monitoring capability. This in-
cludes a double Dolby system so
that the playback is heard with the
correct frequency response and
noise level while a recording is being
made. Certain operating sequences
of the transport controls will also
switch the monitor outputs from tape
to source (but not from source to
tape). One must keep an eye on the
red MONITOR LED to determine
which program is being heard. The
Dolby NR button inserts and re-
moves the Dolby noise-reduction cir-
cuits. There is also a headphone
jack with its own independent
PHONES level control.

Barely visible on the front surface
of the deck are the two microphone
jacks and two pushbutton switches.
One button is for selecting the micro-
phone sensitivity to accommodate
high- and low-output microphones
(the microphones can have bal-
anced or unbalanced outputs and
any impedance from 100 to 5000
ohms). The other button is a TEST
switch that controls an internal
400-Hz oscillator used both for re-
cord head azimuth adjustment and
Dolby-level adjustment.

Like the few other true three-head
 cassette decks with physically sepa-
rate recording and playback heads,
the Model CCD provides for an azi-
muth adjustment on its recording
head. This permits one to compen-
sate for any skew effects in the cas-
sette that could alter the azimuth re-
relationship between the tape and
head. (The playback head is fixed,
and is factory aligned against a
standard test tape.) This record-
head alignment is required to obtain
the full high-frequency response of
which the deck is capable, although
Eumig states that "for most record-
ings this adjustment will probably not
need to be made." The other manu-
facturers of three-head cassette
decks are emphatic about the impor-
tance of making this adjustment for
every cassette before recording and
even when recording on the second
side of a cassette after the deck has
been adjusted for the first side.

To make the alignment, the deck is
put into the record mode and the
TEST button is engaged. In general,
the green LEDs on the level display

Product Focus

Many, if not most, cassette recorders
use servo-controlled dc motors for their
capstan drive. They have the advantage
over synchronous motors of being unaf-
fected by line-frequency changes and
are readily adjustable in speed and re-
tatively inexpensive.

Most such motors use a tachometer
generator of some sort on the motor shaft
to generate an ac output voltage whose
amplitude and frequency are proportional
to the speed of rotation. Either voltage or
frequency can be used to generate the
correcting voltage that drives the capstan
motor. By comparison against a stable
reference voltage or frequency, the ser-
vo action maintains the motor at a con-
stant average speed, but as a rule the
mass of the motor is too great for the
driving amplifier to correct for short-term
fluctuations that can create flutter. A
heavy flywheel, generally on the capstan
shaft, is used to smooth out these fluc-
tuations. The appreciable amount of time
required for the flywheel to reach its op-
erating speed demands that it be left
turning, even while the deck is in PAUSE.
To put the tape into motion, a rubber
pressure roller is used to hold the tape
against the rotating capstan.

In the Model CCD deck, Eumig has
done things quite differently. The motor is
a very-low-mass "coreless" motor type.
It is inherently able to respond to input volt-
age changes much faster than more
massive motors. Also, there is no fly-
wheel: in its place, on the capstan's shaft
is a very-low-mass optical disc on which
there are 2500 radial spokes. At normal
tape speed, the spokes make a beat be-

tween a LED and a pho-
tocell to generate a 15,000-Hz signal.
This signal is compared with a reference
(it is not made clear whether the compari-
son is on a frequency basis or it is first
changed to a dc voltage) and the error

gain is amplified to drive the motor.

Instead of depending on the mass of a
flywheel to eliminate flutter from the tape
drive, Eumig actively eliminates it at the
source by a very-fast-acting motor speed
servo that can smooth out the drive pul-
sations at the capstan. The motor is able
to respond in only milliseconds. The re-


Laboratory Measurements.
The operating manual that accom-
panies the Model CCD makes no
recommendations or suggestions
about specific types of tape to be
used with the deck. (It is, however,
sufficiently complete in its descrip-
tion of the deck's operation and fea-
tures.) An individually run frequency-
response curve is supplied, also
without reference to the tape or even
the recording levels, making it of lit-
te value. We took advantage of the
desk's monitoring capability to test
some 16 tapes in an effort to find the
one (or ones) that best matched the
bias and equalization settings.

Among the ferric-oxide tapes, we
tested BASF Professional I, Fuji
FX-I, Maxell UD-XL I, Maxell LN,
Memorex MRX3, Scotch Dyna-
range, Scotch Master I, and TDK
AD. It was obvious that the ma-

chine's bias was too low for the high-
performs tapes in this group, all
of which had a more or less rising
end-response at a -20-dB re-
cording level. However, Scotch Dynarange and Maxell LN yielded virtually flat response curves, with the slightly better high-frequency saturation of the Scotch tape making it our choice for the balance of the tests with Fe tape.

Among ferrichrome tapes there is much less choice. We tested BASF Professional II, Scotch Master III, and Sony FeCr. The deck was clearly not suitable for use with the first two, which had severe high-frequency losses. The Sony tape provided an acceptable, though not particularly good response compared to the other tape types. This is the tape we used for our FeCr tests.

As expected, the cobalt-treated ferric "chrome-equivalent" tapes gave the best results with the Model CCD. We tested BASF Professional II, Fuji FX-II, Maxell UD-XL II, Scotch Master II, and TDK SA. Although all gave acceptable results, the TDK SA had an impressively flat response that left no doubt that it was the best tape for this deck as it was biased and equalized.

The frequency response at -20 dB with TDK SA was flat within ±0.5 dB from 47 to beyond 20,000 Hz. The head "ripples" at low frequencies had little effect down to about 45 Hz, but the output dropped about 5 dB at 35 Hz and below, which applies to any of the tapes we used. At the high end, the response was down 3 dB at just beyond 21,000 Hz. The 0-dB record/playback response displayed the expected saturation effect. It dropped off beyond 2000 Hz and intersected the -20-dB curve at 10,500 Hz.

The Sony FeCr tape gave a response within +0.5/-3 dB from 38 to 17,000 Hz at -20 dB. However, its saturation at high levels was more pronounced than we are accustomed to seeing, even on cassette machines. The 0-dB response dropped steeply beyond 1000 Hz and intersected the -20-dB curve at 6300 Hz.

The response with Scotch Dynarange was nearly as flat, though not quite as extended, as the TDK SA response. At -20 dB, it was within ±0.5 dB from 46 to 18,000 Hz, and the 0-dB response intersected the -20-dB curve at 10,000 Hz.

The playback equalization could be measured only for the 120-μs (Fe) response since our standard test tapes do not have the notches required for switching the deck to 70μS and there is no override feature. The low-frequency portion of the response from the TDK AC-337 test tape was very flat. However, beyond 1000 Hz, the output dropped off slightly to about -3 dB at 12,500 Hz. Overall, it was within ±1.5 dB from 40 to 12,500 Hz, which is respectable response for any cassette recorder.

The 19-kHz multiplex filter began to affect the frequency response as low as 10,000 Hz and less, but it decreased it by only about 8 dB at 19,000 Hz. Although the attenuation appeared to increase at higher frequencies, it was relatively low at the critical 19,000-Hz pilot carrier frequency.

The Dolby tracking of the recording and playback circuits was excellent. At -20 dB, the frequency re-
THE SPARKOMATIC SOUND.
CAR STEREO FOR THE TRAVELIN' MAN WHO IS IN TOUCH WITH THE CHANGING TIMES.

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CIRCLE NO. 56 ON FREE INFORMATION CARD
Frequency responses at 0 and -20 dB for three different tape types

To produce a 0-dB recording level indication on the LEDs, a line input of 50 mV was needed. The microphone sensitivities were 0.068 and 0.77 mV for the two positions of the microphone attenuator. The corresponding overload levels were 7.7 and 52 mV. The playback output voltage, which is fixed, was 0.46 volt for FeCr tape, 0.57 volt for Scotch Dynarange, and 0.66 volt for TDK SA, all from a 0-dB recording input at 1000 Hz. A comparison with the playback output from a TDK Dolby-level cassette revealed that the playback “Dolby level” of 200 nW/m corresponded to a recording made at -3 dB on the deck’s own amplitude display, but there was no Dolby mark on the display scale, nor was any reference made to this matter in the manual.

The playback waveform from a 0-dB recording with any of the tapes was severely distorted. It contained from 5.6% to 7.1% of third-harmonic distortion. To obtain the usual refer-

The deck’s measured (0.035%) flutter actually surpassed the remarkable claims made for it!

Performance Specifications

<table>
<thead>
<tr>
<th>Specification</th>
<th>Rating</th>
<th>Measured</th>
</tr>
</thead>
<tbody>
<tr>
<td>Tape speed</td>
<td>1 1/4 ips ± 1%</td>
<td>1 1/4 ips +0.15%</td>
</tr>
<tr>
<td>Frequency response (tapes not specified)</td>
<td>Fe: 30-16,000 Hz ±3 dB</td>
<td>46-18,000 Hz ±0.5 dB</td>
</tr>
<tr>
<td></td>
<td>CrO₂: 20-20,000 Hz ±3 dB</td>
<td>47-20,000 Hz ±0.5 dB</td>
</tr>
<tr>
<td></td>
<td>FeCr: 20-20,000 Hz ±3 dB</td>
<td>38-17,000 Hz ±0.5/-3 dB</td>
</tr>
<tr>
<td>S/N (A-weighted)</td>
<td>Dolby Off</td>
<td>Unwtd</td>
</tr>
<tr>
<td></td>
<td>Fe</td>
<td>CCIR/ARM</td>
</tr>
<tr>
<td></td>
<td>Cr</td>
<td></td>
</tr>
<tr>
<td></td>
<td>FeCr</td>
<td></td>
</tr>
<tr>
<td>Bias frequency</td>
<td>175 kHz</td>
<td>Not Checked</td>
</tr>
<tr>
<td>Wow &amp; flutter</td>
<td>0.05% wrms</td>
<td>0.035% wrms</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0.05% wpk (DIN)</td>
</tr>
<tr>
<td>Rewind time (C-60)</td>
<td>40 seconds</td>
<td>47 seconds</td>
</tr>
<tr>
<td>Fast forward time (C-60)</td>
<td>NA</td>
<td>84 seconds</td>
</tr>
<tr>
<td>Start-up time</td>
<td>Less than 0.04 s</td>
<td>Not checked</td>
</tr>
<tr>
<td>Mixing control range</td>
<td>-65 to +15 dB</td>
<td></td>
</tr>
<tr>
<td>Alc range/response time</td>
<td>40/0.01 s</td>
<td>Not checked</td>
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<tr>
<td>Output level/impedance</td>
<td>500 mV/10k ohms</td>
<td>Level OK (depends on tape)</td>
</tr>
<tr>
<td>Headphone output</td>
<td>8-2000 ohms, 1/4&quot; jack</td>
<td></td>
</tr>
</tbody>
</table>
obtained in a combined record/playback measurement. The speed was 0.15% fast and did not change as a cassette was being played. The crosstalk from left to right channel at 1000 Hz, measured with a TDK AC-352 tape, was a relatively high -33 dB, which is still adequate for a full stereo effect.

The Model CCD is claimed to have a very fast rewind speed, and it did. It rewound a C-60 cassette in only 47 seconds, which is appreciably faster than the 80 or more seconds of a typical cassette transport. However, its fast forward was slower, taking 84 seconds to move the same amount of tape. Headphone volume through 200-ohm phones was excellent. The headphone amplifier built into this recorder makes it one of the very few cassette decks that can actually produce a loud listening level with just about any kind of dynamic phones.

**User Comment.** Functionally and operationally, this appears to be a well-conceived and executed deck. Its rather unusual features are easy to get used to; afterward, it is difficult to imagine a recorder smoother or easier to use. The solenoid-operated transport is astonishingly silent. We heard none of the thumps or clunks usually associated with solenoid operation. The tape heads move up to contact the tape with a "motor" rather than a linear-solenoid sound. This is apparently not far from the actual case, since one of the very few details supplied about the inner workings of the deck in the Eumig literature is the fact that the head assembly is pivoted and moves through an arc to contact the tape. However this is accomplished, the heads appear to contact the tape firmly but slowly, with a minimum of impact. The record azimuth adjustment was easy to make and not very critical, unless one is concerned with the response at the highest audio frequencies (beyond 15,000 Hz).

Eumig also claims a start-up time of less than 40 ms for the low-mass capstan motor so that the pause start and stop action (which actually shuts off the motor, instead of leaving it running with the pressure roller released, as is usually done) occurs instantly and with no audible wow. We confirmed this in use. In general, we were most impressed with the mechanical performance of the recorder.

We are less enthusiastic about the relatively low saturation level of the recording head. With most tapes, the 0-dB record/playback frequency response was not as wide as we have found on many other cassette decks in the upper-price range. Even the very measurable difference in response between the -20- and -40-dB recording levels with TDK SA tape suggests that the deck should be operated with a very low average recording level, which needlessly restricts its dynamic range. This is why we obtained S/N figures that, though very acceptable, are not quite as good as expected from a $1300 cassette recorder.

True, the noise levels are low, but so is the maximum recording level.

When recording from FM broadcasts, the deck did a "perfect" job in the sense that operating the monitor button to listen to the incoming program or the tape playback resulted in absolutely no audible difference in the sound. When we tried recording FM tuner interstation hiss at various levels and making this comparison, we could always hear some difference in the extreme highs. The playback was usually slightly dulled, even at levels as low as -20 dB. With high-energy tapes such as TDK AD, which had a rising high-end response, the effect was the reverse, with the playback sounding slightly bright. In both cases the difference was slight, probably too small to detect when recording FM broadcasts or dubbing from most records.

Our conclusions are, of course, based on tests with just one sample of the Model CCD. Hence, we have no way of knowing how representative our test deck was of the company's entire production. If the Model CCD could be given just a few decibels of added headroom in recording, it would be unarguably one of the finest decks on the market at any price. This deck is ingeniously designed, ruggedly made, and highly versatile, with basically excellent performance—when using a limited number of suitable tapes and only when the maximum recording level is kept to -3 dB or less as read on the LED indicators.

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The Pioneer Model TVX-9500 TV sound tuner is designed to take advantage of the wider bandwidth available now to TV networks for intercity audio transmission. This makes it possible for the audio enthusiast to obtain high-fidelity sound from TV programs. It is a compact accessory that contains separate vhf and uhf TV tuners, a high-quality FM i-f and quadrature detector, and audio outputs that can drive a high-level input of any hi-fi system. The TVX-9500 measures 16½" W x 13¼" D x 3¾" H (42 x 35 x 9.8 cm) and weighs roughly 11½ lb (5.2 kg). Suggested retail price is $250.

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**Pioneer TVX-9500 TV Sound Tuner**

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**General Description.** The TVX-9500 resembles other Pioneer audio components in style. Across most of the width of the panel are 12 flat, mechanically interlocked pushbuttons that respond to a very light pressure and select the vhf TV channels from 2 to 13. Above each button is a red LED that glows when that channel is selected. Slightly to the right of the channel 13 button is a uhf button, which transfers tuner operation to the conventional rotary-detented uhf channel selector. The scale of the uhf channel selector is calibrated from 14 to 83, visible through a small window on the panel. It has a concentric fine-tuning ring. A small green LED, labelled TUNED, glows only when a signal of sufficient strength is tuned. On the rear apron are separate antenna inputs for uhf and vhf antennas (both 300-ohm balanced and 75-ohm unbalanced for vhf and 300-ohm balanced for uhf). Even though the audio output is monophonic, there are two audio output jacks to simplify connecting the TV audio to both channels of a stereo amplifier or receiver. Accessible through the bottom panel are a number of small control shafts, one for each vhf channel, that can be adjusted by hand or with a screwdriver. They are for tuning to the sound carrier frequency on each channel. As each is correctly tuned, the green TUNED LED on the front panel comes on. (While making this adjustment, a slide switch on the bottom panel must be set to defeat the AFC and muting circuits. When all channels have been tuned correctly, the AFC and muting are put back into service.) The audio outputs of the tuner are normally silent until a sufficiently strong signal has been received.

**the TVX-9500... dramatically upgrades TV sound**

The performance specifications of the tuner are comparable to those of a good stereo FM tuner. They include a 65-dB signal-to-noise ratio (based on the 25-kHz maximum rated deviation of a TV-sound channel); 50-dB quieting sensitivity of 32 dBf (22 µV); and frequency response of 50 to 10,000 Hz. +0.5/-1 dB. Distortion is rated differently in two places in the brochure that describes the tuner, at either 0.07% or 0.13%.

Relatively simple circuits are used in this tuner, compared to those used in most stereo-FM tuners. The independent vhf and uhf "front ends" appear to be conventional TV tuners. A single-transistor i-f amplifier and a ceramic filter are followed by an integrated circuit that performs limiting and quadrature detection. A low-distortion audio amplifier, powered by bipolar supplies, delivers a nominal 400-mV output to a 4700-ohm load when the input signal has ±25-kHz deviation.

**Laboratory Measurements.** We tested the TVX-9500 on vhf Channel 6, whose sound was within the tuning range of our Sound Technology FM signal generator. The muting threshold was 35 dBf, or 30 µV (rated 34 dBf, or 28 µV). At this signal level, the S/N was already 60 dB. We did not attempt to verify the 50-dB quieting sensitivity rating, since it fell below the muting threshold level.

Minimum distortion was reached at a 45-dBf (100-µV) input and remained constant at higher inputs. The THD + N was 0.14% with a 25-kHz deviated signal, and the unweighted S/N was 64 dB (excellent tuner performance, when one considers that the 75-kHz deviation used in FM broadcasting would have added 10 dB to this figure). Capture ratio, rated at 1.0 dB, was not measured, owing to the muting action of the tuner. In any case, it could not be a significant specification, since cochannel or multipath interference of sufficient magnitude to interfere with sound quality would certainly result in a useless picture from the TV receiver.

The AM rejection improved from 52 dB to 58 dB as the signal level was increased from 45 to 75 dBf (rated at 50 dB at an unspecified input level). The audio frequency response was within +0.15/-0.5 dB from 50 to 15,000 Hz and was down 2 dB at 30 Hz. We noted that the tuner sound cut off when the modulation frequency exceeded 18,000 Hz. Perhaps the muting circuit interprets this high frequency as a noise signal, normally associated with insufficient signal strength. Audio output with 25-kHz deviation was 400 mV, exactly as rated.

**User Comment.** We have been playing our TV sound through a hi-fi system for more than 20 years (deriving it from the detector output to bypass the TV audio system entirely). The result was sometimes much better than the normal TV sound, but more often was merely a more faithful reproduction of intercarrier "buzz" from our old TV receiver. Of course, the limited bandwidth of TV-network shows over most of that period made any improvement marginal at best.

When we used the TVX-9500, the difference was most dramatic. The quality was comparable in every way to that of most FM broadcast stations. This may have been due, in part, to the limitations of the broadcast material as well as to the improved TV sound standards and the TVX-9500. The point is that the sound was of full "hi-fi" quality and listening to it alone would give one no inkling that it came from a TV broadcast.

This is not a "sensitive" tuner in the sense that we use when talking about FM-tuner performance. It is more comparable to a TV receiver in its signal requirements, which is perfectly reasonable. Pioneer suggests connecting it to the TV antenna through a signal splitter to divide the signal between the picture and sound receivers. Alternatively, a separate antenna can be used for the TVX-9500. One thing that is not likely to prove satisfactory is a simple indoor dipole or "rabbit-ears" antenna. Even if the latter seems to give a satisfactory picture, it is unlikely that one could sacrifice half its output to share with the sound tuner without degrading the performance of the picture.

TV-sound tuners have been manufactured before, of course. As we recall, they were rudimentary devices that could not be called "hi-fi" tuners. Furthermore, the TV program quality of that time made even their minuscule improvement unnecessary. Today things are different, and until such time as TV receivers with built-in high-quality audio systems become available (which may never happen), the TVX-9500 provides a very practical way to dramatically up-grade TV sound. We use the term "dramatically" intentionally, since there is nothing subtle about the improvement. Incidentally, if the speakers of a system are located on either side of the TV receiver, the sound will appear to emerge from the screen.
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The Upcoming New World of TV Reception

How information carried in the vertical interval can be used for a variety of personal and commercial purposes

Most people are unaware of many things occurring on their TV receiver's screen because they're normally invisible. For example, a V.I.R. (vertical interval reference) signal is being transmitted by some TV broadcasters to permit receivers with V.I.R. circuitry to electronically compensate for signal variations that would otherwise deteriorate the quality of TV color intensity and tint. You don't see it on the TV screen, just as you don't see other pulses, but it's there nevertheless!

Besides these transmitted test and control signals, there are a number of unseen, experimental transmissions taking place—captions for the deaf, hundreds of pages (screenful) of alphanumeric data and simple graphics that cover weather forecasts, financial news, etc. Home video terminals are gaining in importance rapidly, using a TV signal's vertical blanking intervals. Experimental systems of this type are already working in Japan, Germany, and Great Britain, so we're not spearheading anything truly new in this field. Indeed, the Electronic Industries Association (EIA) is trying to determine if U.S. broadcasters can utilize foreign technical standards for alphanumeric and graphics consumer TV communications.

In addition to the use being made of the vertical blanking lines of TV signals, computerized video/audio advances are making it feasible to develop a TV color "light pen" (see photo above), and TV facsimile machines are using stereo sound broadcasting (the latter an actuality in Japan, but not yet in the United States). With so much going on, then, it is interesting to examine some of the workings of these next-generation video electronic marvels.

Using Special Signals. Over the years, the FCC has authorized special signals to be "piggy-backed" on normal TV-program transmissions. These signals must not interfere with the normal program's picture or sound, of course.

There are four areas where such signals can be inserted. They include the horizontal blanking interval; the vertical blanking interval; the audio channel, using time and/or frequency multiplexing; and the video channel (which also employs time and/or frequency multiplexing). Of these possible options, research revealed that the best possibilities for accepting special signals are in the horizontal or vertical blanking intervals because they occur outside of the nominal screen viewing area (most TV receivers are overscanned horizontally and vertically). Hence, any special signals will not be visible on the screen, even if they modulate into white.

To understand why the vertical interval was chosen as the best spot for adding signals, let's examine the nature of both the horizontal and vertical intervals—the former first.

Horizontal Interval. The horizontal interval, illustrated in Fig. 1, consists of several portions, the most outstanding of which is the sync tip that synchronizes
the receiver's horizontal sweep circuits. Sync intervals are 63.5 µs apart (reciprocal of the 15,734-Hz horizontal transmission frequency). Note the reference signal levels. Maximum carrier level at the sync tip is the maximum output power of the transmitter. Slightly below the maximum output is the reference black level. Anything greater in amplitude than this reference is "blacker-than-black" and will therefore be invisible. The blanking level is established in this non-visible zone.

The reference white level represents a low transmitter output power. A picture video signal can span the full range from black to white by swinging between the black and white reference levels. Zero carrier is never attained, since at that level the transmitter has no output.

A narrow "front-porch" leading element in the pulse forms the black framing bar seen at the right of the viewed image and is usually off-screen. Since this portion is a narrow 1.27 µs in duration, not very much can be inserted in the way of special signals.

The sync pulse (about 5 µs wide) synchronizes the receiver's horizontal circuit. The tips of the pulse appear to be a good place to insert special signals, but this could impose difficulties with sync-separation circuits. So it, too, cannot be used to accommodate new signals.

The "back porch" (about 5 µs wide) forms the left-side black framing bar of the viewed image, which is also usually off-screen. This location has already been allocated to the color burst consisting of eight cycles of the 3.58-MHz color signal. Adding another signal here could cause reception problems, so this area, too, is not a good prospect for inserting special signals.

**Vertical Interval.** The vertical interval was chosen to carry special signals owing to its long duration. Here's how it evolves. Each horizontal sweep "paints" a narrow line across the CRT screen. When it gets to the right side, the electron beam is blanked (made invisible) and the horizontal circuit causes the beam to rapidly retrace to the left side to continue the cycle. This retrace takes less than 10 µs. The much slower vertical sweep causes the glowing line to move down a little for each successive horizontal sweep. This action produces the picture raster. But, what happens when the horizontal sweep reaches the bottom of the picture? Here is where the vertical blanking interval comes in.

At this point, internal blanking circuits cut-off the CRT electron beam so that it does not produce a glowing line on the CRT, and the vertical sweep circuit causes the now-dark beam to retrace to the top of the screen. The transmitted sync signals are arranged so that this invisible vertical retrace takes the time equivalent of 21 horizontal lines. Since each horizontal line is 63.5 µs long, the vertical interval occupies a little more than 1333 µs. But what is more important, it contains 21 horizontal lines that

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**Fig. 1.** The horizontal interval as used for color. Color burst amplitude is same as sync tip height.

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**Fig. 2.** There are 21 horizontal intervals between start of a field and its actual video.
are not used for the viewed image.

Also, engineering studies have shown that a smoother visual display will occur if one frame of video is divided into two fields, with one interlacing the other.

To avoid picture "tearing" (unsynchronized for several horizontal lines), horizontal sync pulses, or their equivalents in time, must be transmitted during the vertical interval. Also, since 525 horizontal scanning lines were specified for the NTSC (National Television System Committee) signal, each of the two viewed fields had to contain 262.5 lines. It was decided to produce the half line just before the start of field two (Fig. 2). Thus, one field begins with a complete horizontal line having its origin at the top left corner of the screen and ending with a half line at the bottom of the screen. The next field starts with a half line that has its origin at the top center of the screen and ends with a whole line at the bottom.

Note that all activity in the vertical-blanking interval occurs in the blacker-than-black area so that it is not visible. All synchronization occurs on the positive-going edge of a pulse.

The first nine lines of a field are occupied by six (pre) equalizing pulses, the vertical-sync pulse, and six (post) equalizing pulses that occur at half-line intervals and are half the width of horizontal-sync pulses.

During vertical retrace, the receiver's horizontal oscillator is locked to its frequency by the equalizing pulses. Without equalizing pulses, the horizontal oscillator may drift far enough off-frequency to require several lines to get it back into sync, causing the picture to "tear." Because they are at twice the horizontal frequency, the equalizing pulses permit the half-line start at the beginning of field two.

Vertical sync consists of one long pulse of three times the horizontal interval time, which is broken up into six sections by inverted equalizing pulses. The positive-going edges of alternate sections provide the sync for the horizontal oscillator to assure that it maintains lock during the vertical sync interval. After the second set of equalizing pulses has passed, the signal reverts to conventional horizontal sync (with color burst) until the video starts again.

There are, then, a total of 21 real and equivalent horizontal lines from the beginning of the field to the start of that field's video. The first nine lines have been described, leaving lines 10 through 21 available for other use.

Back in 1967, RCA conducted a test of special signals on lines 10 through 17 for its "Homefax" system. The system used vertical-interval signals to produce facsimile copy on a printer attached to a TV receiver. Tests were also conducted in Japan, involving lines 14 through 21 and digital data transmission.

An expanded view of the currently used lines 17 through 21 is shown in Fig. 3. There are some similarities and some

(Continued on page 55)
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differences between signals on both fields. It is possible to see these signals if you adjust the vertical hold of your TV receiver until the thick black bar rolls down into view. It may be difficult to make this bar stationary, but it can be kept in place long enough to see the signals. The vertical-interval signals are the one-scanning-line-thick dot-and-bar structures that can be seen on the five scanning lines immediately above the picture, at the bottom of the black bar. Some of these components may be in color. Bear in mind that not all stations carry the same vertical-interval test signals (VITS), and non-network stations may not have any.

Lines 17 and 18 carry the VITS. When provided by a network, these signals are used to check network transmission facilities. Provided by a local station, they are used to check transmitter performance. These signals essentially emulate a complete video and chroma test signal that is continuously in operation, even when the normal TV picture is being displayed. The signals do not interfere with the on-screen picture. Station technicians simply pick them off for making measurements.

Line 18 of both fields currently carries a "staircase" of the 3.57-MHz bursts (the chroma reference) at progressively lighter shadings for differential phase and gain checks of a transmitter. The two sine-squared pulses are used to make one-pulse checks of the system, since a single sine-squared pulse contains all the frequencies the system must pass. The larger of the two pulses modulates the chroma reference. These two pulses are followed by a white "window" that checks for transient ringing.

Signals on lines 17 and 18 are required to be radiated by remote-controlled TV transmitters. These lines are used for network transmission of test signals, which are different in format and which would be deleted prior to radiation by the remote-controlled station.

Line 20 of field one is currently an experimental line. Some network stations carry a source identification code during this interval.

Line 21 is used for picture captioning, a procedure used to help the estimated 13.4-million hearing-impaired people in the U.S. In 1975, Public Broadcasting System (PBS) began an "open" caption format for some of its news programs. It is currently being carried by 125 member stations.

As shown in Fig. 4, all of line 21 field one and the first half of line 21 field two are used for captioning purposes. The data signal is a non-return-to-zero (NRZ) format that uses standard 7-bit plus parity ASCII coding. The framing code used by the associated decoder is transmitted during the first half of line 21 field two, when data reference and test signals are present. The center waveform in Fig. 4 shows the reference pulse, transmitted every eighth frame, to be used by the decoder-associated multipath equalizing filter.

In operation, caption data is removed from the vertical interval and displayed at the bottom of the screen. The present PBS system is capable of transmitting written text at faster than 550 words/minute. It is expected that caption-decoding equipment will be ready for use to serve the hearing-impaired public in the not too distant future.

**Vertical Interval Reference.** The basic difference between the VIT (vertical interval test) and the VIR (vertical interval reference) signals is in applica-
tion. VIT signals are used to monitor transmission equipment and system performance, while VIR monitors the parameters of the color program being transmitted.

The VIR, shown in Fig. 5, occupies line 19 of both fields. The signal consists of a horizontal-sync pulse, color burst, chroma reference (same frequency as burst), luminance reference, and black reference.

Different color-TV manufacturers use different approaches to using VIR signals. The method used in General Electric's VIR-II system is shown in Fig. 6. When the chrominance and black reference levels are equal in amplitude at the receiver's R-Y output, chroma phase (color tint) conforms to that of the transmitted reference signal. When these two signals are equal at the blue drive output, the chroma level (color) is matched to the transmitted reference signal.

Line 19 is selected as shown in Fig. 7. The vertical-sync pulse toggles a bistable flip-flop connected to the reset input of a seven-stage counter. Horizontal pulses are connected to the clock input. The delay allows the flip-flop to toggle on the second serration of the vertical-sync pulse. This allows both fields to toggle the flip-flop during the fourth scan line so that the count always begins at the same point in each field.

There is only one state where four of the outputs from the counter are all 1's simultaneously, that is at line 19. (Binary 1111 equals 15 scanned lines, which when added to the four preset lines, brings the total to 19 lines.)

Outputs from the counter are coupled to a diode AND gate that produces a 63-µs pulse (one horizontal line interval) at exactly line 19. An inhibit line turns off the flip-flop until the next field (vertical) sync pulse. The process then repeats.

The line-19 gate signal goes to a chroma-interval pulse generator where a 20-µs pulse that corresponds to the chroma-reference interval on line 19 is created. When present, this signal generates the operating voltages for the tint- and color-level controllers.

The VIR sensor (Fig. 6) detects the presence or absence of the VIR signal. If VIR is not present, the output voltage of this circuit causes the manual/auto logic to switch to manual operation of the color and tint controls. If it is present, the VIR indicator lamp comes on and the output of the circuit switches into auto so that the receiver itself controls the color and tint.

Since the key to using VIR depends on accurate location of the vertical-sync pulse, some means must be used to ac-
The basic circuit waveform is identified, another circuit locates get through. Once hence, only the higher el waveforms. vary, as different, pulses removed by a capacitor circuit. The composite sync from a conventional sync separator causes Q2 to turn off during the period of these pulses. The horizontal-retrace pulse cuts off Q3 during this interval. When Q2 and Q3 are cut off, the Q4 base-circuit capacitor charges up during the equalizing and vertical-sync interval. When Q2 and Q3 are conducting, the capacitor discharges, keeping the base of Q4 at ground potential. Since the intervals between equalizing and vertical pulses are different, the pulses at the base of Q4 vary, as illustrated by the accompanying waveforms.

The base voltage of Q5 is set to a level about halfway between E1 and E2. Hence, only the higher-amplitude pulses get through. Once the vertical is clearly identified, another circuit locates line 19.

A pulse-width-to-height converter is used by Matsushita to remove the vertical sync from the composite waveform. The basic circuit and its associated waveforms are shown in Fig. 9. The transistor turns on and supplies charging current to \( C_b \) through \( R_b \) when its base is high. When the base goes low, at time t1, the transistor cuts off and \( C_b \) discharges through \( R_b \) and \( D_1 \). The voltage across \( R_b \) decreases according to the \( C_b R_b \) time constant.

When a high is applied to the base at time t2, the transistor turns on and instantly raises the voltage across \( R_b \) and \( R_a \). During time t2 to t3, a charging current to \( C_b \) is produced via \( R_b \). Depending on the amount of charging current, a gradually decreasing voltage, determined by time constant \( C_b R_b \), appears across \( R_a \).

At time t3, the transistor cuts off. This discharges \( C_b \) and decreases the voltage across \( R_a \). This voltage decrease depends on the cutoff time of the transistor. The pulse height across \( R_a \) is determined by the width of the vertical pulse segments. The pulses that appear across \( R_a \) are sent to a threshold circuit that allows only the vertical-sync pulses to pass. Once derived, the clean vertical sync operates a line-19 counter.
Teletext. The term "Teletext" is used to describe a system for transmission of alphanumeric characters and simple graphics during the vertical interval. The data is extracted and displayed on-screen in place of conventional video. With such a display system, there is almost no limit to what can be "printed" on-screen—with color capability.

Several countries are using or experimenting with one form or another of Teletext. Only one TV station in the U.S. (KSL in Salt Lake City) is testing a form of Teletext at present.

In England, CEEFAX and Oracle (Optional Reception of Announcements by Coded Line Electronics) systems have been used since 1974. CEEFAX is also being tested in Sweden, Australia, and New Zealand. Canada is experimenting with VIDEOTEXT, while France is using ANTIOPÉ. Sweden has tested EXTRATEXT for picture subtitling. Italy and Bavaria have tested digital encoding.

Japan is trying out a couple of systems. In one, both Japanese and other ideographic symbols can be transmitted. In the other, still pictures are transmitted a couple of lines at a time (within the vertical interval). The lines are stored on a magnetic medium until a full frame is present and then played back in real time. Even music can be digitized.

KSL's Teletext and CEEFAX transmitted signals are similar (Fig. 10). American Teletext is licensed by the British, which explains the similarity. However, some differences between the two systems exist due to the different scanning rates in the two systems. Data transmission is via the NRZ technique, which allows reduced bandwidth. During encoding, discrete pulses are not used. Instead, data is transmitted as 0 (low) and 1 (high) voltage levels. If a series of 1's is transmitted, the signal remains high for all adjacent 1's instead of dropping to zero between each pulse. For multiple
0's, the signal simply remains low for all adjacent 0's. A clock signal determines whether a 1 or a 0 occurs at clock time.

Since all transmission paths are subject to distortion-inducing errors, particularly multipath in TV reception, code-protection schemes are used. The character code is 7-bit ASCII, with odd parity for a total of eight bits. The address code uses four parity bits in a Hamming code.

The clock's 16 run-in bits are used to lock the decoder clock in frequency and phase in a similar fashion to that used to lock the receiver's chroma oscillator to the color burst in a color receiver. The framing code indicates the start of the first 8-bit word; the 16 bits following the framing code (address) identify which of the 20 rows is being transmitted. These signals are also protected by the Hamming code.

The page data identifies the page being received, while the hours and minutes bytes are used to insert the local time in the displayed image. Alternatively, they can be used to preset the system to accept data transmitted at a certain time. Remaining control bits perform system "housekeeping," while the rest of the line contains the actual data.

U.S. Teletex has 20 rows of 31 characters/row (UK system uses 23 rows of 40 characters) and operates at seven pages/second. ASCII characters are transmitted with a choice of six colors, including white. Special video effects—flashing, boxing, and inversion—can be performed using nonvisible control characters. Up to 800 pages of "magazine" can be transmitted, using the 5.5-megabaud rate in lines 15 and 16. When this rate is spread over the entire TV screen, it drops to about 30 kilobaud. The pages are transmitted in a repeating cycle with any necessary updating.

In the Salt Lake City tests, the receivers used the Texas Instruments developed TIFAX decoders. These decoders contain a signal slicer that removes the pertinent lines from the two fields; one page of RAM; a character generator;
color matrixing; and various timing and control logic. All subcircuits are implemented on LSI chips.

The Teletext keypad has four keys. One is used for conventional TV viewing; the second "calls" the Teletext and deletes the conventional video; the third allows for mixed TV and Teletext (Teletext in monochrome here); and the fourth is used for page selection. In use, the viewer selects a page number via the keypad. The page number appears at the upper left corner of the screen. A special "header" line displays the various pages available alongside the requested page header, which remains stationary. When the selected page header is received and decoded, the data it contains fills the screen. This page remains on-screen until it is deleted or another page request is made.

Telecomputing. Since the Teletext system is a digital CRT alphanumeric/graphics display (in color if desired), it operates like a "dumb" computer terminal. Even a dumb terminal, however, can be made "smarter" by adding a microprocessor and support logic.

Engineers at England's ORACLE have, in fact, designed and built such a system, using the logic shown in Fig. 11. Note the similarity to any other small computer in that there is a bus system, CPU, some RAM, an operating system (control program) in ROM, and several I/O ports. What sets this system apart is that it gets its software via the Teletext system.

This "Telesoftware" is transmitted as conventional Teletext lines and is selected by the user from a "menu" that appears on one regular page. If a high-level language like BASIC is stored in ROM, there is a very wide variety of available programs. Programs can be longer than one Teletext page, since two or more pages can be merged. The number of programs available depends on how many lines the service allows before interference with the regular pages. It may well be that at some future time, other vertical-interval lines will be used just for computer programs. In addition to home computing, this system can be used for small business computing or to play arcade-type games.

Once the service is running and a bus has been established, there is almost no end to what peripherals can be plugged in. This will make the home TV receiver the ultimate video system in that it will interact with the viewer. Perhaps one day, when cable facilities are adequate, two-way transmissions will come into being.

Teleview. General Instrument Corporation's version of a Teletext receiver is shown in block form in Fig. 12. This "Teleview" approach is similar to the British approach and looks like a basic computer. It uses a microprocessor-based controller that also contains an "operating system" in built-in ROM, RAM to store data, a video generator to drive the CRT color and luminance circuits, and a data acquisition module that accepts the Teletext data removed from the vertical interval by the Data Grabber.

Provisions are provided to accept optional telephone-line data (Viewdata) via a modem. Unlike conventional computers, the keyboard can be directly wired to the controller or operated via an optional infrared link. Other I/O ports to make hard copy and cassette recordings are in the works. At this writing, these circuits employ the GI-fabricated data acquisition chip No. G0977-11, video generator No. G0977-12, and No. PIC1650 microprocessor-based controller No. G0977-13. The infrared circuit uses AY-3-8471 and AY-3-8475 trans-
mitter and receiver chips, respectively. The telephone-line transmission unit is soon to be released, as are other I/O port devices.

The single-chip PIC1650 microcomputer contains 32 8-bit registers, 512 x 12 bit ROM, ALU, four sets of eight user-defined TTL-compatible I/O lines, real-time counter, and clock and uses a single 5-volt line. Both keyboard scanning and display can be performed simultaneously. This chip controls the address and data lines and uses eight bits for the keyboard, eight bits for the data bus, and sixteen bits for address and control signals.

The video generator defines which TV line is expected to contain Teletext data and sends a signal to the Data Grabber to accept that line just before the Teletext's clock run-in is expected. The received Teletext data is then decoded and is checked for any transmission errors by the data acquisition module.

Pages of text are requested through the keyboard and controller. Teletext pages are identified by a three-digit code—such as 100 to 199, 200 to 299, etc.—called a "magazine." At present, page numbering goes to 899 (eight magazines). When a page is requested, its number appears at the upper left of the display. As the various pages "roll through," they are momentarily displayed until the requested page comes up. When the page number matches the selected code, the data is "dumped" into RAM for storage and display. Any page can be erased or updated as desired, and up to eight different pages can be stored and accessed without delay.

In the picture mode, the normal TV image is seen on-screen and any special newsflash and subtitle pages appear in a box in the TV picture. Characters in other pages can be displayed by operating a REVEAL/CONCEAL switch.

In the mix mode, the TV picture is displayed and the incoming Teletext data appears in monochrome characters superimposed on the screen. Another user-selectable function includes an on-screen "clock box" to display time.

The Viewdata setup receives data via telephone lines. Because reception is asynchronous with the Teleview system, a separate data path is required. Up to three characters can be received during a TV frame. Tests are still being made on the telephone-line option. Although no information is available, we can assume that other semiconductor manufacturers are developing their own versions of Teletext chips.

As a result of using phone lines, the capability of two-way data exchanges is inborn. For example, a subscriber could have a dialogue with a computer data bank to, say, call up his bank statement.
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Fun and Teach. At the most recent Consumer Electronics Show, January 1979, Matsushita Electric Co. (parent for Panasonic and Quasar) demonstrated some exciting video products that were not commercially available. One was called the "Fun and Teach" machine. It used an ordinary color TV receiver in conjunction with a light pen to “write” or “draw” directly on the TV screen (see lead photo). Furthermore, recordings were made for later playback on a standard audio cassette machine.

As shown in Fig. 13, the home education-amusement video system section is essentially a computer. It contains a microprocessor, 2K of RAM, an operating system in ROM, separate video-display RAM, a cassette I/O port, and a combined keyboard/light-pen I/O. Elements are interconnected by a combined address/data/handshake bus.

The video display features 96 vertical by 128 horizontal dots for graphics and 192 V x 256 H dots for alphanumeric character display. Seven colors are available: red, green, blue, yellow, cyan, magenta, and white. Displayed data can be stored on standard C-60 cassettes.

The system’s stereo cassette recorder utilizes one channel for the data, the other for narration and music. I/O for the cassette is via a UART, with serial-to-parallel inputting from tape and parallel-to-serial when recording. The system speed is 4800 baud.

In use, a seven-color “palette” is displayed along the bottom of the screen (see front-cover illustration) when the light-pen program is selected. Touched to the desired color and then placed against the screen, the light pen permits the user to draw in the selected color. Colors can be changed as desired. With a little care, a series of frames can be drawn and recorded (along with suitable narration) so that basic animation with it’s own sound track can be created.

A selection of cassettes was prepared at the show. One contained a fairy tale in still color pictures and narration. Another was an "etch-a-sketch" type of program. Others contain foreign languages, with the text appearing on screen as a teaching aid. Also, the operating system permits the user to create his own tapes, using both on-screen data and voice and/or music.

Kindergartens and homes where there are small children would be good targets for the Fun and Teach machine. Children will be able to view nursery tales and narration and picture-book presentations and will also be able to draw on the screen in colors of their own choosing. Drawings can be stored on ordinary audio cassettes for later viewing.

The alphanumeric mode allows the usual home-computer data to be entered, displayed, and stored as desired. Thus, games and animation sequences can be created, accompanied by sound effects. Up to 350 video frames can be stored on a standard C-60 cassette.

Another Matsushita-developed working model consisted of a facsimile read-out machine attached to a color TV receiver. Printout of alphanumericics and graphics were accomplished in about one minute from a separately transmitted program using one of the stereo TV sound channels as authorized in Japan. The implication here is that one can get printouts at any time of news, etc., without interrupting TV viewing and listening in the process.

Clearly, new TV applications have been developed that promise to be implemented on a wide public basis in the near future. EIA subcommittees have already been formed to develop recommended broadcast standards for both multichannel television sound and Teletext data and graphics transmission. Moreover, the IEEE (Institute of Electrical and Electronic Engineers) has a summer meeting scheduled to explore the many facets of promised upcoming consumer TV uses. And according to a report, "The Home Terminal," by International Resource Development, Inc., New Canaan, CT, interactive TV such as the Qube experiment in Ohio and Viewdata in England, point to a strong consumer demand for these TV services. IRD predicts that this type of TV home terminal will appear on the U.S. market in 1982, with a price of $1400, enabling users to pay bills and enjoy electronic mail services by video transmission techniques.
Perform Complete Impedance Measurements with This R-F Bridge

By Don Morar, W3QVZ

Inexpensive bridge measures $R$ and $X$ components over a wide frequency range.

One of the most useful instruments an experimenter who works with r-f circuits can have is an impedance bridge. The ideal bridge would permit accurate measurement of both the resistive and reactive components of an unknown impedance over a wide range of frequencies. Commercial r-f impedance bridges, although they satisfy these requirements, are priced well beyond the means of the average experimenter. On the other hand, those affordable bridges that have appeared as construction projects in amateur radio and hobby electronics magazines only tell half the story—the resistive component.

The bridge described in this article can measure the complex impedance of just about any load at frequencies between 3.5 and 54 MHz with a high degree of accuracy. Moreover, it can be inexpensively built using “junk box” components, and is smaller and lighter than its typical commercial counterpart. The only external items required for calibration and operation are a group of nonreactive resistors, an r-f source such as a signal generator, and, of course, the impedance to be measured.

Among the project’s features are a built-in null indicator (a microammeter) and an amplifier which can be switched into the null detector circuit to enhance its sensitivity. The value of the impedance’s resistive component is read directly off the bridge’s $R$ dial, which is calibrated in ohms. The unit’s $X$ (reactance) dial calibration is scaled in terms of frequency. This is done because inductive and capacitive reactance vary with frequency, so an $X$ dial calibrated directly in ohms would be accurate at only one specific frequency. Scaling the $X$ dial’s calibration in terms of frequency provides greater operating flexibility.

About the Circuit. The schematic diagram of a basic r-f bridge is shown in Fig. 1. It obviously resembles the classic Wheatstone bridge, which has four resistive arms. Two of these arms are usually derived from a potentiometer. The r-f bridge, however, employs a dual differential variable capacitor (C1C2) so that measurements can be performed over a wide range of frequencies. If a potentiometer were used, its frequency-dependent intrinsic reactance would cause the bridge to yield false results.

Some readers might not be familiar with the dual differential capacitor. It is, essentially, two variable capacitors ganged so that when one section (capacitor) exhibits maximum capacitance, the other exhibits minimum capacitance. The use of a dual differential capacitor provides a variable capacitance ratio between the two arms of the bridge that it comprises. This simplifies the calibration and use of the bridge. If the value of $R_1$ is a constant, the single control knob used to vary the setting of C1C2 can be calibrated in terms of $R_1$ or in ohms.

To use the bridge the unknown impedance is connected as $R_2$ and an r-f source is used to energize the network. The dual differential capacitor is then adjusted so that the bridge is balanced. When that happens, no voltage drop will exist across the bridge detector and no current will flow through it. The detector will indicate a null and the value of $R_2$ can be read off the dual differential capacitor’s control knob.

The bridge in Fig. 1 will only measure the unknown’s resistive component. More complete information about the unknown (including its reactance) can be obtained by measuring it on the bridge shown in Fig. 2. This circuit re-

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Fig. 1. Basic r-f bridge uses dual differential capacitor but gives only resistive information.

Fig. 2. More sophisticated bridge can measure reactance as well as resistances.

MAY 1979
sembles that of Fig. 1, but two components (C3 and L1) have been added to the bridge's lower right arm. To underscore the bridge's greater measuring facilities, the unknown is no longer represented as a resistance (R_u), but as a general impedance (Z_u).

As in Fig. 1, dual differential capacitor C1C2 is used to measure the real (resis-
tive) component of the unknown complex impedance. Variable inductor L1 and variable capacitor C3 make possible measurement of both the sign and magnitude of the unknown's imaginary (reactive) component. The bridge thereby provides the user with complete information about the unknown impedance.

The device is initially balanced at the frequency of interest with a purely resistive termination at Z_u. Variable capacitor C3 is placed at its midrange setting and inductor L1 is adjusted for resonance. This cancels out any reactance which would otherwise be reflected into the other bridge arms. The nonreactive termination is then replaced with the unknown impedance. Its resistive component is balanced by varying C1C2 and its magnitude read off the calibrated capacitor control knob scale. The unknown's imaginary component is balanced by shifting C3 away from its mid-scale setting in either the clockwise (+, the standard sign for inductive reactance) or counterclockwise (−, the standard sign for capacitive reactance) direction to cancel out any reactance in the unknown.

If the unknown impedance has an inductive component, more capacitive reactance (that is, less capacitance) is required from C3 to obtain a balance. Conversely, if the load has a capacitive component, more capacitance is needed. With both capacitive and less capacitive reactance is required. Once C3 has been properly adjusted, the bottom right leg of the bridge will look purely resistive, and an excellent null will be obtained on the detector.

The scale of C3's control knob should be calibrated in terms of the magnitude and sign of the reactance present at Z_u. Because inductive reactance varies directly with frequency and capacitive reactance varies inversely with frequency, the calibration of C3's control knob must be scaled in terms of frequency. If it were calibrated directly in ohms, its calibration would hold true at one frequency only. A better approach is to perform the calibration at 1 MHz and frequency-scaled it. The exact magnitude of the reactance component can then be determined by a simple arithmetic operation.

The complete schematic of the r-f impedance bridge is shown in Fig. 3. Resistive balancing is performed by the dual differential capacitor C1. Noninductive resistor R2 provides the reference against which the resistive component of the unknown impedance is measured. (The unknown is connected to LOAD jack J3.) Balancing and measurement of the unknown's capacitive component (if any) is the task of L2 and C2.

When the bridge is unbalanced, germanium diode D1 rectifies r-f into pulsating dc which is filtered by L1 and C3. If S1 is placed in its direct position, the filtered dc is applied to null indicator M1, a 0-to-200-μA meter. For increased reso-

**PARTS LIST**

| B1 | 9-volt transistor battery |
| C1 | Dual differential variable capacitor, 12- to-150 pF per section, Millen No. 28801 or equivalent (see text) |
| C2 | 14.7-to-339-pF variable capacitor (Millen No. 19335 or equivalent) |
| C3 | 0.01-μF disc ceramic capacitor |
| D1 | 1N82 or equivalent germanium diode |
| D2 | SO-10 2N5458 amplifier (see text) |
| J1 | Standard Amphenol 4-prong jack |
| J2 | Millen 1100 (see text and Table 1) |
| J3 | SO-239 coaxial connector |
| R1 | 2N5458 n-channel JFET |
| R2 | 4-prong plug to match J2 |
| Q1 | 2N5458 n-channel JFET |

Misc. — 10" × 6" × 3 1/2" (25.4 × 15.2 × 8.9 cm) aluminum utility box (Bud CU301A or equivalent), 5/4" × 3" × 2 1/2" (13.4 × 7.6 × 5.4 cm) aluminum utility box (Bud CU3006A or equivalent), J.W. Miller No. 42000CBI or equivalent slug-tuned coil forms, plastic or metal threaded BX/Romex outdoor electrical-box plugs, 1/2-inch (3.8 cm) PVC pipe, PVC pipe adapters (1/8-inch or 1.3-cm threads to 1/4-inch or 1.7-cm pipe), cyanoacrylate cement, control knobs with 1/4-inch (3.3-mm) shaft hole, one small control knob with 1/4-inch (6.6-mm) shaft hole, two large Bakelite control knobs with 1/4-inch (6.6-mm) shaft hole threaded porcelain standoffs, one noninsulated and two insulated 1/4-inch (6.6-mm) shaft couplings, L brackets, battery clip, battery holder, perforated board, several nonreactive resistors whose values have been accurately determined, PL-259 coaxial connectors, convenient lengths of 50-ohm coaxial cable, enameled copper wire, hook-up wire, solder lugs, solder, machine and self-rapping hardware, etc.
olution of the null, $S_1$ should be placed in its amplified position. The filtered dc is then amplified by $Q_1$, which in turn drives the meter movement. Use of the amplifier also increases bridge sensitivity so that the circuit is compatible with low-level signal sources such as solid-state "grid" dippers.

Construction. In any r-f bridge, it is essential that residual and stray reactances be kept to a minimum, and this project is no exception to that rule. Placement of components must be such that lead lengths in the r-f portion of the circuit are absolutely as small as possible. The layout established by the author, which can be seen in the photograph of his prototype (Fig. 4), yielded good results up to 54 MHz.

All of the components were mounted in an aluminum utility box measuring 10" x 6" x 3½" (25.4 x 15.2 x 8.9 cm). The frames and stators of variable capacitors $C_1$ and $C_2$ must be insulated from ground (the enclosure), necessitating the use of threaded porcelain spacers or their equivalent. Similarly, insulated couplings should be used with the capacitors' control shafts.

Dual differential capacitor $C_1$ is partially hidden in the photograph by one half of an aluminum utility box which mounts inside the main enclosure and shields the capacitor from the rest of the bridge. (The other half of the utility box has been removed to expose the capacitor for the photograph.) Dimensions of the box shield used by the author are 5½" x 3" x 2½" (13.4 x 7.6 x 5.4 cm). Totally enclosing the differential capacitor within the grounded utility box helps keep stray reactances small.

To cover 3.5 through 54 MHz with one variable capacitor ($C_2$) requires the use of several different inductors. However, band switching of the inductors is not used in this project because it would introduce too much stray reactance and degrade bridge performance. The author's solution to this problem is to use plug-in inductors. J.W. Miller coil forms (No. 42000CBI), ½-inch (1.3-cm) inner-diameter PVC pipe fittings, and 4-prong plugs are used in making the coils.

Details of coil construction are shown in Fig. 5. First, the various coils should be wound on slug-tuned forms. Coil winding data appears in the Table. After the coils are wound, they should be soldered to standard Amphenol four-prong plugs. (Bases removed from discarded four-prong vacuum tubes can be used instead of four-pin plugs.) Take the suggested PVC pipe fittings and modify them as shown in Fig. 5. Then affix each four-pin plug to a modified PVC fitting with cyanoacrylate cement (Eastman 910, "Krazy Glue," or equivalent).

Using a ¼-inch (6.5-mm) bit, drill out the center of a ½-inch (1.3-cm) plastic or metal threaded BX/Romex outdoor electrical-box plug to accommodate the Miller coil form's metal bushing. Mount the threaded plug in the PVC fitting and attach a knob to the coil form's tuning shaft to complete the coil assembly. Repeat this procedure for each inductor.

Fig. 4. Photo of author's prototype. Note small shielding box (half of which has been removed) partially obscuring dual differential variable capacitor.

Fig. 5. Assembly details for each plug-in coil ($L_2$).
Either a commercial dual differential capacitor or a home-brew one can be used for C1. The dual differential capacitor should have a capacitance of 12-to-150 pF per section. A Millen No. 28801 dual differential capacitor is suitable, but the author ganged two identical Hammarlund receiving-type variable capacitors rated at 12-to-150 pF each. If two capacitors are used, they should be ganged so that one is at maximum capacitance (plates fully meshed) when the other is at minimum capacitance (plates fully open). The other variable capacitor, C2, is rated at 14.7-to-339 pF. A Millen should be installed between the two rotor shafts.

Other details of the construction of the author's prototype are apparent in Fig. 4. A portion of the small shield box has been cut away with a nibbling tool to provide room for J1, R1, R2 and the lead connected to the rotor plates of C1. The null detector's amplifier is mounted on a small piece of perforated board which is mechanically supported by L brackets secured to the terminals of Mt. Because parts placement is critical in the r-f portion of the project, it is best to duplicate the author's layout closely.

Calibration. The resistance (r) dial of the impedance bridge can be calibrated by using an assortment of 1/2-watt carbon composition resistors of various values within the 5-to-200-ohm range. The author measured the exact resistance of each component selected on a General Radio GR-650 bridge to enhance the accuracy of the calibration. If you don't have access to a highly accurate bridge, measure the resistors with a good-quality digital multimeter or use close-tolerance metal film components. Connect the resistors to PL-259 coaxial plugs, keeping lead lengths short.

Calibration should be performed at 3.5 MHz to minimize the effects of reactive strays. Apply the output of a signal generator oscillating at that frequency to J1 and connect the first load resistor (the one with the lowest resistance) to J3. Also, install the 80-meter plug-in coil at J2. With S1 in its DIRECT position and C2 set at 50% dial rotation (plates half meshed), adjust C1 and L1 for the best null possible. Then place S1 in its AMPLIFIED position and fine-tune for the deepest null you can obtain. Place a notch, tick mark, or other notation of the position on C1's dial. Repeat this procedure for each calibrating resistor.

The author used Bakelite knobs with large skirts as the x and r control knobs. Calibration of the r knob was made by inscribing the appropriate point on the Bakelite skirt with tick marks and numerical values using an electric engraving tool. This technique permits direct calibration of the r knob in ohms. (Suitable Bakelite knobs are available from such surplus electronics dealers as Fair Radio Sales Co., Box 1105, Lima, OH 45802.) Alternatively, a knob with a silver skirt calibrated from 0 to 100 over 180 degrees of dial rotation can be used in conjunction with a graph of dial readings plotted against resistance values.

No direct calibration was performed on the x (reactance) dial. Rather, the following procedure was used. Using a Southwest Technical Products digital capacitance meter cross-checked against a General Radio GR-650 bridge, the author made a plot of the capacitance of C2 against dial rotation. Then the standard inductive and capacitive reactance formulas were employed to derive a plot of reactance below and above a resonant frequency of 1 MHz. Assuming that L2 is adjusted to cancel out bridge reactance (including that of C2 when its plates are half meshed), the graph shown in Fig. 6 plots the net reactive variation of Xc and XL below and above the resonance at 1 MHz.

This graph can be used to calibrate the x control knob. For example, at 50% dial rotation, the reactance of the load is 0. At 75% dial rotation, the reactance is ±740 or 740 ohms inductive. Similarly, at 25% dial rotation, the reactance is

![Fig. 6. Calibration curve for the X (reactancescale of C2's control knob.)](image-url)
would depart curate. How much calibration depends on the graph of Fig. 6 to determine the sign and magnitude of the reactive component.

The accuracy of the x control knob's calibration depends on that of the graph of Fig. 6 and the degree of bridge balance (null sharpness) obtainable. The theoretical curve is apparently very accurate. How much a direct calibration would depart from the curve would depend on stray bridge reactance. The prototype yielded good, sharp nulls and its x calibration was very accurate.

Using the Bridge. Before an unknown impedance can be measured, it is necessary to balance the bridge. Apply an r-f signal to J1 and connect a non-reactive termination to J3. (The author employs a commercial 50-ohm, 5-watt nonreactive termination when performing this step.) Any signal source producing 1 to 3 volts rms of r-f can be used. A grid-dip oscillator loosely link-coupled to J1 is satisfactory. The author employs a 4-turn coil of No. 16 enameled copper wire large enough to accommodate the outer diameter of his grid-dip coil to apply r-f for 80- and 40-meter measurements and 2 turns of the same wire for measurements on 20, 15, 10, and 6 meters. Each coil is connected to a convenient length of 50-ohm coaxial cable, the other end of which is terminated with a PL-259 connector.

Plug the appropriate coil for the frequency at which the measurement is to be performed into jack J2. Then set the x control knob to 0 ohms (50% rotation or midscale). Adjust C1 and L2 for a good null as indicated by M1. After initial adjustments, switch the amplifier into the meter circuit to increase the resolution of the null. If a complete null cannot be obtained, reduce the coupling between the signal source and the bridge.

After the bridge has been balanced, replace the purely resistive load with the unknown impedance. Alternately adjust C1 and C2 to obtain the best null and note the readings of the r and x scales. Impedance measurements are in rectangular form. An impedance with an inductive component is of the form \( Z = R + jX \), where \( R \) and \( X \) are the readings of the \( r \) and \( x \) scales, respectively. The operator +j denotes inductive reactance, and \( F \) is the frequency at which the measurement is performed. An impedance with a capacitive component is of the form \( Z = R - jX / F \), where \( R \), \( X \), and \( F \) are as defined in the case of a partially inductive impedance. The operator \(-j\) denotes capacitive reactance.

As mentioned earlier, the x measurement involves frequency scaling. In the case of an inductive reactance, the exact magnitude is determined by multiplying the x scale reading by the frequency at which the measurement is performed. The exact magnitude of a capacitive reactance can be obtained by dividing the x scale reading by the frequency at which the measurement is made.

In Conclusion. Here are a few hints that you should keep in mind when using this project. Bridge measurements are of course frequency sensitive. The bridge must therefore be rebalanced after a frequency change of 1% or more occurs. Be sure to balance the bridge with a purely resistive test load before performing any measurements. Otherwise, inherent bridge reactances will cause a false reading.

Remember that the bridge requires very little r-f drive. This is no problem when a signal generator or grid-dip oscillator is used as the signal source because the output level of the generator or the coupling between the oscillator and the bridge can be easily reduced. However, if a transmitter is used to provide r-f for the impedance measurement of, say, an antenna or linear amplifier input stage, care must be taken not to overload the bridge. The transmitter's r-f output must be kept at a low level, and the bridge must not be left in the line when more than 0.1 watt of r-f power is flowing.

It is usually very inconvenient to perform impedance measurements directly at an antenna's feed point, so they are commonly performed at the transmitter end of the transmission line. This can result in misleading information if the line is not an integral multiple of an electrical half-wavelength. Note that a transmission line's electrical length is its physical length expressed in free-space wavelengths at the frequency of interest multiplied by the line's velocity factor. Solid-dielectric coax (RG-58, RG-59, RG-8, RG-11, etc.) has a velocity factor of approximately 0.66; polyfoam coax has a velocity factor of approximately 0.81.

If it is not convenient to add or subtract enough cable to make the transmission line an integral multiple of an electrical half-wavelength, a Smith chart can be used to transpose the measured impedance at the transmitter end of the line into the actual antenna impedance. To do this, the line length must be accurately determined by physical measurement or by measuring it with a grid-dip oscillator and the far end of the line shorted. Remember that you must employ the electrical length of the line when using the Smith chart.

You are now ready to start using your impedance bridge in r-f work. Its usefulness on your test bench or in your radio shack will be quickly appreciated.

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**COIL WINDING DATA**

<table>
<thead>
<tr>
<th>Band</th>
<th>Approximate Frequency Range</th>
<th>Coil Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>80 M</td>
<td>3.4 to 4.2 MHz</td>
<td>26 turns of No. 30 enameled wire, close wound</td>
</tr>
<tr>
<td>40 M</td>
<td>6.5 to 7.5 MHz</td>
<td>16 turns of No. 22 enameled wire, close wound</td>
</tr>
<tr>
<td>20 M</td>
<td>13.0 to 15.0 MHz</td>
<td>8 turns of No. 16 enameled wire, close wound</td>
</tr>
<tr>
<td>15 M</td>
<td>19.5 to 22.0 MHz</td>
<td>3 1/2 turns of No. 16 enameled wire, close wound</td>
</tr>
<tr>
<td>10 M</td>
<td>27.0 to 30.0 MHz</td>
<td>21 1/2 turns of No. 16 enameled wire, close wound</td>
</tr>
<tr>
<td>6 M</td>
<td>50.0 to 54.0 MHz</td>
<td>1 turn of No. 16 enameled wire</td>
</tr>
</tbody>
</table>

All coils are to be wound on a J.W. Miller No. 42000CB1 or equivalent slug-tuned form.

MAY 1979
Open-Door "Fridge Alarm" Stops Food Spoilage & Energy Waste
REFRIGERATORS are among the hungriest of household appliances in terms of electrical power consumption. Every time a refrigerator door is opened, cold air spills out and the warm air that replaces it must be cooled. Needless to say, it pays in dollars and cents to limit the time the door is open to as brief a period as possible. The low-cost Fridge Alarm described here maybe just what you need to limit the time you study the contents of your refrigerator or your child forgets to close the door.

The Fridge Alarm is a photoelectric device that is activated as soon as the door opens and the refrigerator's light goes on. It sounds an insistent two-tone signal if the door remains open past a given number of seconds.

### About the Circuit

As shown in Fig. 1, when light strikes its photosensitive surface, Q1 triggers into conduction and causes Q2 to saturate. This places pin 1 of IC1 close to ground potential and allows the timer to start operating (Fig. 2). Since the voltage across C1 is initially zero, IC1 is triggered into immediate operation. Timing is controlled by R8, R1, and C1.

During the timing sequence, the output of IC1 at pin 3 remains high (almost at VCC) and keeps IC2 and IC3 cut off, since pin 1 of each of these integrated circuits is connected to this line.

Most electrolytic and many aluminum capacitors can have sizable leakage currents. Hence, they should not be used in timing circuits. To avoid this problem, C1 should be a tantalum capacitor. Using the time constants shown, R8 can be set for periods of from 4 to 17 seconds. (This range was selected because 8 seconds is about the mean time for access to a refrigerator.) Because C1 discharges through D1 and the 15,000-ohm internal resistance of IC1, pin 7 is left unconnected.

If the light striking Q1 is interrupted during the timing cycle, both Q1 and Q2 turn off and timing capacitor C1 rapidly discharges through D1 and IC1, resetting the timer. In darkness, Q1 has a very high collector-emitter resistance. With Q2 in cutoff, standby current is extremely low.

Should the light striking Q1 be constant, the timing cycle will run its course and the output at pin 3 of IC1 goes low. This effectively grounds pin 1 of both IC2 and IC3, activating these ICs.

Integrated circuits IC2 and IC3 are wired to operate as astable multivibrators. The oscillating frequency of IC2 is about 4 Hz. This 4-Hz signal "modulates" IC3, and the output of IC3 directly drives a small loudspeaker.

The two-tone sound is created by alternately shunting the IC2 end of R4 between VCC and ground at a 4-Hz rate. When pin 3 of IC2 is high, the parallel combination of R4 and R5 produces about a 500-Hz tone. When pin 3 is low, R4 is effectively shunted to ground. This reduces the voltage at pin 7 of IC3. Since C6 must now charge to 80% and then discharge to 40% of this new value to activate the comparators inside IC3, about a 330-Hz tone is generated. The two tones alternate at a 4-Hz rate as long as the circuit is activated.

### Construction

All components, except B1 (and its optional battery holder) and the small loudspeaker can be mounted on a printed circuit board. The actual-size etching-and-drilling guide and components-placement guide for the printed circuit board are shown in Fig. 3.

The leads of Q1 can be identified with the aid of an ohmmeter and light source if an unmarked phototransistor is used.

The project can be mounted inside a small translucent box that permits sufficient light to pass through and trigger Q1 into conduction. Any of the various polyethylene refrigerator-type storage containers on the market will suffice as long as they are large enough to accommodate the circuit. The loudspeaker is best secured to the bottom of the container (after drilling a number of small holes for

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**Sounds an alarm after preset time when refrigerator door is left open**

*BY ELLIOT K. RAND*

MAY 1979
FRIDGE ALARM

Continued

Fig. 1. Timing action of circuit is initiated by light striking Q1.

Fig. 2. Block diagram of principal circuits in the 555 IC. In this case, one 555 is used as timer, and two as astable multivibrators.

Fig. 3. Components are mounted on board as shown at left and enclosed in a translucent box.

PARTS LIST

<table>
<thead>
<tr>
<th>Component</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>B1</td>
<td>9-volt battery</td>
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<tr>
<td>C1</td>
<td>10-µF, 25-V tantalum capacitor</td>
</tr>
<tr>
<td>C2, C5, C7</td>
<td>0.01-µF disc capacitor</td>
</tr>
<tr>
<td>R1</td>
<td>300,000 ohms</td>
</tr>
<tr>
<td>R2, R5</td>
<td>10,000 ohms</td>
</tr>
<tr>
<td>R3</td>
<td>6.8 megohms</td>
</tr>
<tr>
<td>R4</td>
<td>47.0 ohms</td>
</tr>
<tr>
<td>R6</td>
<td>24,000 ohms</td>
</tr>
<tr>
<td>R7</td>
<td>180,000 ohms</td>
</tr>
<tr>
<td>R8</td>
<td>1-megohm trimmer potentiometer</td>
</tr>
<tr>
<td>AP</td>
<td>Miniature 8-ohm loudspeaker</td>
</tr>
<tr>
<td>Misc.</td>
<td>Battery holder, translucent plastic refrigerator container (about 3&quot; square); silicone-rubber cement; hookup wire; etc.</td>
</tr>
</tbody>
</table>

Note: The following items are available from Rand Laboratories, P.O. Box 468, Cape Canaveral, FL 32920: complete kit of parts including drilled case for $9.95 postpaid. Also available: pc board only, $4.25 postpaid. Florida residents, please add sales tax.

the sound to escape down through the shelf with silicone-rubber cement. The speaker and pc board are interconnected with #20 wire so that the board can be positioned to allow maximum exposure of Q1 to the lamp.

The assembled alarm can be tested by placing it in a darkened location and shining a light on it. After a several-second delay, the alarm should sound. Count the number of seconds between the time the light goes on and the alarm sounds. Adjust R8 as needed for the desired delay between the two events.

Place the Fridge Alarm inside your refrigerator in a location where it will receive the maximum amount of light from the refrigerator's lamp. Make sure it is in a location where there will be no possibility of liquid spills on it. Equally important, make sure that the selected location will obliterate any possibility of obstructing the light.

C3, C4, C6—0.05-µF disc capacitor
C8—10-µF, 25-V aluminum capacitor
D1—IN4148 or similar diode
IC1, IC2, IC3—555 timer
Q1—FPT100 or equivalent
Q2—2N2222 or similar transistor
All resistors 1/4-watt, 10% tolerance:
R1—300,000 ohms
R2, R5—10,000 ohms
R3—6.8 megohms
R4—47.0 ohms
R6—24,000 ohms
R7—180,000 ohms
R8—1-megohm trimmer potentiometer
AP—Miniature 8-ohm loudspeaker
Misc.—Battery holder, translucent plastic refrigerator container (about 3" square); silicone-rubber cement; hookup wire; etc.
MOTORIZED bikes—mopeds—are becoming increasingly popular as laws prohibiting them are lifted. However, their low speed and, generally, young drivers, combine with the absence of lighted turn signal to cause safety problems.

Adding the turn-signal system described here will likely reduce accidents. It features a solid-state circuit, low-power lamps, can be used with a moped's 6-V negative-ground electrical system, and can be built for less than $10. It can also be modified for use with a bicycle if a 6-V battery is added.

Circuit Operation. As shown in the diagram, a 555 timer (IC1) is used to
generate 1-Hz pulses. Note that this circuit will not operate until its pin 1 is connected to ground. When turn-select switch S1 is placed in either its L (left) or R (right) position, this ground connection is made through either D2 or D3, depending on the switch position selected. When one of these states occurs, IC1 cycles at its 1-Hz rate, opening and closing relay K1. Relay contacts direct power to the selected lamps (I1 and I2 for left front and rear, I3 and I4 for right front and rear).

Ground for selected lamps is made through the switching member of S1. Thus, as long as S1 is at one of its turn positions, the selected lamps will glow in 1-Hz cycles. When S1 is placed in its center (off) position, the circuit is isolated and stops working.

Zener diode D4 and capacitor C3 maintain a smooth 6-V dc if the output of the motor-powered generator varies.

Construction. The circuit can be assembled on a small piece of perforated board, or a small printed circuit board can be designed.

The circuit can be mounted in a small plastic box that can be secured to the moped frame. The circuit ground should be made to a good metal connection on the moped frame, while the 6 volts is taken from a source that is live when the ignition key is turned on.

The rear turn-signal lamps are inexpensive “bullet-lamp” assemblies that use two GE #81 single-contact bayonet lamps, while the front turn-signal lamps are made from an AMF “rear directional light.” Both of these are obtainable from most bicycle or discount shops. The AMF unit comes with turn-switch, lamps, case and battery holder. Lamps are replaced with GE #27 lamps. The battery holder is not used, and the wiring between the holder and lamps is removed.

The rear lamps must be insulated from the frame. The easiest way to do this is to mount them on the plastic rear license-plate holder. If you mount them to the metal frame, use some form of insulation between the lamp bracket and the metal surface.

Turn switch S1 can be any large bat-handle switch having good detents and a center-off position. It can be mounted as desired on the handlebar.

Interconnections between the switch, power, lamps and the electronic circuit should be made with well-insulated flexible wire passed through a length of plastic tubing taped to the frame.

Photo shows how rear indicator lamps are insulated from moped frame by mounting them on the plastic license-plate holder. Connections to other parts of system must be through insulated flexible wire.

**Fig. 1.** When pin 1 of IC1 is connected to ground through S1, the 555 timer generates 1-Hz pulses to open and close relay K1.

**PARTS LIST**

- C1—10-µF, 15-volt electrolytic
- C2—0.01-µF, 15-volt disc
- C3—1000-µF, 15-volt electrolytic
- D1, D2, D3—IN4001 or similar diode
- D4—6-volt 1-ampere zener diode
- F1—1-ampere in-line fuse
- IC1—555 timer
- K1—6-volt relay
- S1—3-position large bat handle switch (center off)
- Misc.—Plastic enclosure, perforated board, wiring, plastic tubing, mounting hardware, #27 lamps.
Tape Bias/Equalization Chart

By Craig Stark

Many stereo cassette decks nowadays have switches to set bias and equalization for each tape type. But tape formulations come and go, so keeping track of what tapes require which settings can be a challenge. To make recording and playback easier for you, listed below are most of the major high-fidelity cassette formulations of the past few years. A line divides the current and discontinued tape formulations in each manufacturer's product lineup. The discontinued tapes are listed for the benefit of those who still have some of these cassettes on hand.

<table>
<thead>
<tr>
<th>Cassette</th>
<th>Ferric</th>
<th>CrO₂</th>
<th>Ferrichrome</th>
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| Note: In each company listing, those tapes below the line are older, discontinued formulations; the others above are either current or near-current (i.e. may be still on some dealer shelves). On Ferric tapes, those which can profit by a slightly higher-than-usual ("Japanese") bias are identified by an asterisk. On CrO₂-type tapes, those identified with an asterisk are not CrO₂ tapes, but so-called "70-microsecond ferrics," usually modified by cobalt.
# Microcomputer Video Board Buying Directory

<table>
<thead>
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<th>Make &amp; Model</th>
<th>Price ($)</th>
<th>Power Required (mA)</th>
<th>ASCII Char set</th>
<th>Lines</th>
<th>Graphics</th>
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<td>128</td>
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<td>24</td>
<td>BC, RV, AC, S; block-mode edit, KB interface, port-addressed, no driver software needed.</td>
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<td>1K refresh memory; upper-case only; char set includes graphics chart; all standard screen formats exc. 80x24; 2K memory, all formats.</td>
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</tr>
<tr>
<td>VIO-AC</td>
<td>60 (k)</td>
<td></td>
<td>96</td>
<td>80</td>
<td>24 or 120</td>
<td>Converts VIO-B to VIO-C.</td>
</tr>
<tr>
<td>VIO-BC</td>
<td>60 (k)</td>
<td></td>
<td>96</td>
<td>80</td>
<td>24 or 120</td>
<td>Converts Basic VIO to VIO-C.</td>
</tr>
<tr>
<td>VIO-CC</td>
<td>150 (k)</td>
<td></td>
<td>96</td>
<td>80</td>
<td>24 or 120</td>
<td></td>
</tr>
<tr>
<td><strong>Interactive Micro Systems</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>IMS64-100</td>
<td>225 (w)</td>
<td>X</td>
<td>128</td>
<td>32</td>
<td>16</td>
<td>*BC, RV, AC, optional 64x64, 16-color graphics, 6802 intelligence.</td>
</tr>
<tr>
<td></td>
<td></td>
<td>X</td>
<td></td>
<td>64</td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>Ithaca Audio</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SVPM</td>
<td>25 (b)</td>
<td>2A</td>
<td>128</td>
<td>32</td>
<td>16</td>
<td>BC, RV, AC.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>64</td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>Jade Computer</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>JG-VB1B</td>
<td>35 (b)</td>
<td>2A</td>
<td>128</td>
<td>32</td>
<td>16</td>
<td>BC, RV, AC.</td>
</tr>
<tr>
<td></td>
<td>100 (k)</td>
<td></td>
<td></td>
<td>32</td>
<td>16</td>
<td>Erase to end-of-line, scroll, Greek chart.</td>
</tr>
<tr>
<td></td>
<td>150 (w)</td>
<td></td>
<td></td>
<td>64</td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>Micro Diversions</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Screen-splitter</td>
<td>329 (k)</td>
<td>1.5</td>
<td>128</td>
<td>86</td>
<td>40</td>
<td>BC, RV, AC; up to 3440 independent text &quot;windows&quot;; APL, Sci. &amp; graphics character sets available; user-programmable char sets.</td>
</tr>
<tr>
<td></td>
<td>429 (w)</td>
<td></td>
<td></td>
<td>96</td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>MSD-Micro Syst. Dev.</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>MSDV-100</td>
<td>285 (k)</td>
<td>600</td>
<td>96</td>
<td>80</td>
<td>24</td>
<td>5x9 matrix for alpha, 6x10 for graphics &amp; connected chars; 32 graphics chars on ROM; gray scale; scrolling register, underline, programmable timers; 2 boards; BC, AV, AC.</td>
</tr>
<tr>
<td></td>
<td>385 (k)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>Polymorphic</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>VTI</td>
<td>210 (k)</td>
<td>1.6</td>
<td>96</td>
<td>32</td>
<td>16</td>
<td>Requires 2.5 MHz CRT bandwidth for 32-char line or 64-cell graphic line, 5.5 MHz for 64 char or 128-cell.</td>
</tr>
<tr>
<td></td>
<td>280 (w)</td>
<td></td>
<td></td>
<td>64</td>
<td>16</td>
<td></td>
</tr>
<tr>
<td><strong>Processor Tech.</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>VDM-1</td>
<td>199 (k)</td>
<td>u/l</td>
<td>64</td>
<td>16</td>
<td>no</td>
<td>BC, RV, AC, S.</td>
</tr>
<tr>
<td></td>
<td>295 (w)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>Solid State Music</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>V81B</td>
<td>150 (k)</td>
<td>1.4</td>
<td>128</td>
<td>64</td>
<td>16</td>
<td>RV, AC; composite &amp; non-composite video.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>VAMP</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Polygrafix</td>
<td>245 (w)</td>
<td></td>
<td>64</td>
<td>16</td>
<td>128-cell</td>
<td>*RV; 128 user-programmable char; piggy-back upgrade for Polymorphics VTI-64 card; $525 w/VTI-64.</td>
</tr>
</tbody>
</table>

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**POPULAR ELECTRONICS**
### Make & Model
<table>
<thead>
<tr>
<th>Make &amp; Model</th>
<th>Price</th>
<th>Power Required</th>
<th>ASCII</th>
<th>Char</th>
<th>Lines</th>
<th>Graphics</th>
<th>Remarks</th>
</tr>
</thead>
<tbody>
<tr>
<td>Vector Graphic Flashwriter</td>
<td>235 ($)</td>
<td>1.2A</td>
<td>+6V</td>
<td>128</td>
<td>64</td>
<td>16</td>
<td>128×48</td>
</tr>
<tr>
<td>Western Data Sys. Pro/Ex 1</td>
<td>296 ($)</td>
<td>2A</td>
<td>125</td>
<td>64</td>
<td>40</td>
<td>24</td>
<td>block</td>
</tr>
</tbody>
</table>

### Make & Model
<table>
<thead>
<tr>
<th>Make &amp; Model</th>
<th>Price</th>
<th>Power Required</th>
<th>ASCII</th>
<th>Char</th>
<th>Lines</th>
<th>Graphics</th>
<th>Remarks</th>
</tr>
</thead>
<tbody>
<tr>
<td>F&amp;D Associates VDB-1</td>
<td>29 ($)</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>64</td>
<td>64</td>
<td>X</td>
</tr>
<tr>
<td>Gimin</td>
<td>198 ($)</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>64</td>
<td>64</td>
<td>16</td>
</tr>
<tr>
<td>Interactive Micro</td>
<td>225 ($)</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>128</td>
<td>32, 64</td>
<td>16</td>
</tr>
<tr>
<td>Xitan VDB</td>
<td>369 ($)</td>
<td>X</td>
<td>96</td>
<td>80</td>
<td>25</td>
<td>180x75</td>
<td>Buffer memory holds two pages; keyboard port; lower-case descenders; 64 graphics chars; BC, RV, AC.</td>
</tr>
<tr>
<td>Xitek SCT-100</td>
<td>157 ($)</td>
<td>X</td>
<td>128</td>
<td>64</td>
<td>16</td>
<td>limited</td>
<td>&quot;Stand-alone, uses only 8V and ground from S-100 bus, interfaces via ports; ASCII or Baudot I/O; char set includes some Greek, graphics chars; part no 395.</td>
</tr>
</tbody>
</table>

1. (b)=bare board; (k)=kit; (w)=wired.
2. in amps where indicated by "A".
3. u=upper case; l=lower case.
4. AC=Addressable Cursor; BC=Blanking Character; RV=Reverse Video; S=Scrolling.

---

**LINE-VOLTAGE COMPENSATOR**

Boosts the power-supply voltage when it drops too low

In areas where low power-line voltage is common, a filament transformer can be used as a voltage booster. A 6.3-volt transformer can be used as shown in the figure. When the switch is placed in the BOOST position, the transformer acts as an autotransformer, increasing the voltage across the socket terminals by about 6 volts. When selecting a filament transformer for this application, determine how much current in amperes the load will draw. Then select a transformer whose secondary winding can safely handle this load current.

The dots shown near the transformer denote phasing of the windings. If you do not know how the transformer is phased, you can determine this experimentally. Connect the secondary wires one way, power the circuit, place the switch in the boost position and measure the voltage across the power socket. If it is higher than the line voltage (the voltage across the primary), the transformer has been wired correctly. If the voltage across the socket is less than the line voltage, reverse the secondary wires. If the transformer has been incorrectly wired with respect to phase, it acts as a "bucking" autotransformer which has a lower output voltage than it has input voltage.

---

**BY HARRY J. MILLER**

With the switch in BOOST, line voltage is raised about 6V.

---

MAY 1979
TWINLEAD CONNECTORS

Surplus crystal holders and sockets with a 1/2" (1.3-cm) pin spacing make excellent connectors for 300-ohm “twinlead” transmission line. If the crystal holder’s front plate is held in place with machine hardware, remove the machine screws and disassemble the crystal holder. If the holder is hermetically sealed, saw off its top. Heat the holder pins with your soldering iron and remove the thin metal plates that “sandwich” the quartz slab. Remove any excess solder plugging up the two pins. Next, strip the twinlead’s insulation so that about 1” (2.5 cm) of each conductor is exposed. Tin the conductors, and slide them into the holder so that their tips protrude from the ends of the pins. Solder the conductors and clip off any excess. Crystal sockets can be employed as antenna jacks to match the plugs you have fashioned. Because there are so many crystals ground for “odd” unusable frequencies available on the surplus market at low cost, you will find these twinlead connectors very economical. You might well have several such crystals in your junk box right now. —Harry J. Miller, Sarasota, Fl.

Personal Computing...

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Only during the National Computer Conference will you have an opportunity to experience personal computing to the fullest. And that’s why the 1979 Personal Computing Festival, June 6-7 in New York’s Sheraton Center Hotel, formerly the America, is different. As a conference within a conference, it will give you the chance to explore the complete spectrum of information processing while concentrating on those aspects of personal computing you won’t want to miss…including equipment, applications, ideas, and new developments that have created excitement throughout the entire computing community.

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Against the backdrop of the prestigious NCC, the Personal Computing Festival has attracted many well-known experts and personalities who will participate in an information-packed technical program and compete for prizes for the best presentations. Join them in exploring applications ranging from use of small business systems and financial analysis to personal networking, new information utilities, and aid to the handicapped.

You will also have ample opportunity to discuss new ideas and novel approaches to shared problems, to find out what to expect in the year ahead, and observe interesting and clever applications demonstrated by the individuals who developed them.

Plan now to take part in a unique personal computing experience at NCC ’79. You can register for the Festival at the Sheraton Center Hotel, 522nd Street between 7th Avenue and Avenue of the Americas, for only $15 which includes your copy of the

NCC ’79 Personal Computing Proceedings. Registrations, excluding the Proceedings, are available at $5 for one day and $5 for all four days. The Proceedings will be available separately at $8. For additional information on NCC ’79, including housing and registration procedures, contact AFIPS, 210 Summit Avenue, Montvale, N.J. 07645; phone 201/391-9810. To obtain information on the special NCC Travel Service call toll-free 800/996-6882.

PC BOARDS FORM PROJECT ENCLOSURE

With the price of quality cabinets continually increasing, it’s possible to spend more for a simple project’s enclosure than for all the components. However, there’s another way to house your projects—use their pc boards as their own enclosures (see photo). If you are making your own pc boards, study the schematic before doing any layout or etching. In most cases, you’ll be able to break the circuit down into several sub-assemblies. For example, the project shown is a two-digit tachometer with four pc boards. The boards accommodate a clock, decade counters, displays, and drivers, and support circuitry. If the circuit does not lend itself to such a functional breakdown, simply parcel it out onto separate board using a minimum amount of jumpers.

If you lay out each stage on its own board, place the ground foil along one edge and the supply voltage foil on the opposite side. It will then be possible to join the various boards together at the appropriate edges to form a “sell” enclosure. Four pc boards can form a box with two open sides, five boards a box with one open side, and six a closed box. Breaking a project down into seven or more boards allows you to make eye-catching geometric shapes.

For an aesthetically pleasing look, use pc boards of various colors and place the component sides on the exterior of the “enclosure.” The bottom board can be supported by four rubber feet on it. —James Temple, Bethpage, NY.
THE ANALOG COMPARATOR

The analog comparator is a circuit that compares an input voltage to a reference voltage and changes the state of its output when the input exceeds the reference. This decision-making ability has many important applications, several of which we will examine here.

A simple analog comparator can be made by using an operational amplifier without a feedback resistor. The role that a feedback resistor usually plays is to pass some of the amplified signal back to the inverting input of the op amp, thus reducing the amplifier's gain. Without the gain limitation imposed by a feedback resistor, the op amp operates at its maximum ("open-loop") gain. A small input voltage will then cause the output of the op amp to change state immediately. The resulting voltage swing is so dramatic that the comparator can be considered a switching circuit.

The operation of a noninverting analog comparator is shown in Fig. 1. A known reference voltage is applied to the comparator's inverting (-) input, and an unknown voltage to its noninverting (+) input. The LED indicates the status of the comparator's output.

In operation, the output of the comparator is at -V when the input voltage is more negative than the reference voltage which in this case is ground. The LED indicates this by glowing. When the input voltage is more positive than the reference voltage, the comparator output switches from -V to +V and the LED is extinguished. Because the reference is ground, a very small positive voltage will trigger the comparator. In both cases, the voltage difference is measured in millivolts.

Comparator Demonstration Circuit. Unless you have previously worked with analog comparators, you will probably want to take a few minutes to breadboard the simple demonstration circuit shown in Fig. 2 before trying any of the circuits that will be described later.

The comparator in this circuit is a 741 op amp without a feedback resistor. A variable input voltage is provided by R1, a potentiometer operated as a voltage divider. Resistors R2 and R3 form a fixed voltage divider that provides a reference at half the supply voltage.

When the input voltage is below the reference voltage, the LED glows to indicate that the comparator's output is low (at ground). The LED switches off to indicate the comparator's output is high (at +9V) as soon as the input voltage exceeds the reference voltage. With the values shown in Figure 2, R1's wiper will be at the center of its rotation when the comparator switches, assuming that R1 is a linear potentiometer.

---

**Fig. 1. Operation of a basic comparator circuit.**

**Fig. 2. Schematic of a demonstration comparator circuit.**
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Sine- to Square-Wave Converter.

One of the simplest applications for a comparator is the sine- to square-wave converter shown in Fig. 3. The reference voltage is ground so the comparator switches its output to its maximum positive value when the sine-wave voltage exceeds ground potential. Similarly, the comparator output switches to its maximum negative value when the sine-wave voltage is at or below ground potential. The result is a square wave with the same period as the sine wave.

Peak Detector. Another simple but useful comparator application is the peak detector. As its name implies, the peak detector retains the maximum amplitude of a fluctuating input voltage for subsequent readout and analysis. Suitable transducers connected to the input of a peak detector permit the determination of such parameters as maximum wind velocity, temperature, light intensity, vehicle speed, and many others.

Figure 4 shows a basic peak detector circuit that you can easily assemble. To understand its operation, assume that \( C1 \) is initially discharged (i.e., the \( \text{RESET} \) switch has been momentarily closed). This means that the reference voltage at the inverting input of the comparator is 0 and that a positive input voltage will immediately switch the output of the comparator to +9 volts. The comparator output will then begin to charge \( C1 \) until the voltage across the capacitor equals the input voltage. As soon as the two voltages are equal, the comparator output immediately drops to ground potential and \( C1 \) stops charging.

If a subsequent input voltage exceeds the charge stored in \( C1 \), the comparator output will again go high and allow \( C1 \) to charge to the new peak voltage. This tracking process ensures that \( C1 \) always retains the peak voltage applied to the input. When you want to track a new

Fig. 3. Comparator as sine-wave to square-wave converter.
(lower) peak voltage, close the RESET switch to discharge C1.

The peak detector circuit is subject to drift because C1 will gradually lose its charge. Diode D1 prevents discharge through the comparator, but discharge can take place through the output circuitry or through the dielectric leakage of the capacitor. For these reasons, it is important to use a low-loss polystyrene or Mylar capacitor for C1 and a high-impedance monitoring circuit.

Last month’s installment of this column described a simple high-input-impedance voltage follower you can use to interface the peak detector to a low-impedance device such as a panel meter or VOM. Without the high-impedance buffer, C1 will quickly discharge when you attempt to measure the voltage across it.

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operate in the noninverting mode. That is, they generate an output identical in polarity to the input voltage. However, a comparator can be operated in the inverting mode by simply reversing the two inputs. This makes possible many additional applications, one of which is called the limit or window comparator.

A window comparator can be made from three-fourths of an LM339 quad comparator as shown in Fig. 5. This chip was the subject of the January 1977 Experimenter's Corner. Unlike the 741, the LM339 is specifically designed to operate with a single-polarity power supply.

In operation, IC1C functions as a noninverting comparator, but IC1A operates as an inverting comparator. Potentiometer R1 and fixed resistors R2 and R3 form a divider chain that delivers slightly different voltages to the two comparators. These voltages define the upper and lower limits of the circuit's switching "window," which can be changed easily by varying R2 and R3.

The output of each comparator in the LM339 is an uncommitted collector. This means two or more outputs can be tied together to achieve a logic OR function without using diodes or a logic gate.

When the input voltage is less positive than IC1C's reference voltage, the output collector of this comparator is low. When the input voltage is more positive than IC1A's reference voltage, its output collector is low. When either output is low, the other is pulled low, causing a LED connected between the two outputs and the positive power supply to glow.

If the input voltage falls in the window region between the two reference voltages, the output of each comparator is high. This will cause a LED connected to the outputs to be darkened.

It's usually desirable for an indicator to light when a desired condition is met. The third comparator in Fig. 5 serves this purpose by inverting the output of the window comparator. The LED then glows only when the input voltage falls within the window region.

An even more useful version of the circuit is shown in Fig. 6. Here, the third comparator is employed as a NAND gate. Three LED's connected to the outputs of all three comparators provide a HIGH WINDOW LOW indication. For best results, use a green LED for the WINDOW indicator and red LED's for the HIGH and LOW indicators. The green LED will glow when the input voltage is within the window. The red LED's will indicate that the input voltage is either above or below the window. The LED's should be
mounted in a vertical row with the high LED on top, the window LED in the middle, and the low LED on the bottom.

If you use three different colors for the LED’s, the circuit will tell you whether the input voltage is above, below, or in the window no matter how the LED’s are mounted. A red LED connected to the output of IC 1A, for instance, would indicate a high voltage. A yellow LED at IC 1C would indicate a low voltage. Finally, a green LED at IC 1D would indicate an input voltage within the window.

Incidentally, the comparator used as a NAND gate in Fig. 6 can be replaced by a conventional TTL 7400 NAND gate. In

Fig. 7 shows one possible configuration suggested by Bill Ciciks of Rockford, IL. Regular readers of this column might recall the moving-dot voltage indicator described in the October 1978 installment. Bill’s circuit requires three extra resistors. If you replace the window comparator shown in Fig. 7 with the circuit shown at the lower right of the page, you will have a solid-state, low-power, accurate window comparator. You can also eliminate the green LED at the output of the window comparator shown in Fig. 7 with the circuit shown at the lower right of the page.

**Solid-State Oscilloscope Update.** The solid-state oscilloscope described in previous columns has resulted in more letters than any other topic covered in “Experimenter’s Corner.” In fact, the moving-dot voltage indicator in Fig. 7 is actually the vertical section of a solid-state scope designed and built by Bill Ciciks.

I’ll have more to say about this and other experimental solid-state scopes in a future column. In the meantime, I would like to hear from other experimenters who have successfully assembled and operated all-solid-state oscilloscopes. Please include a stamped, self-addressed envelope for a reply.

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COMMON ANODE vs. COMMON CATHODE DISPLAYS

Q. A circuit calls for common-cathode LED displays, but the ones I have are the common-anode type. Is there a circuit that will allow me to use common-anode displays with common-cathode circuits, and vice versa? —Chris Iannuzzi, Warrenton, VA.

A. The common-anode display, connected to a chip designed to operate with such a display, is shown at A in the figure. When base current is applied to a driver transistor, it turns on and sinks current for the diode in a given segment. (For simplicity, only one diode per segment is assumed and only two segments are shown. Also, multiplexing details are omitted.) Note that all anodes in the display have a common connection to +VCC. A common-cathode display is shown at B. Base current applied to a segment’s driver transistor causes the device to conduct and source current for the LED. All cathodes in the display have a common connection to ground.

As far as the display-driver IC is concerned, one for common-anode applications sinks current through its output stages, but one for common-cathode displays sources current at its outputs. To use a common-anode driver with a common-cathode display, the circuit shown at C can be used. Pull-up resistor R, whose value should be chosen so that a few milliamperes flow through the driver transistor when base current is applied, is connected to +VCC. When the transistor conducts, the input of the inverter is grounded. Accordingly, the output of the inverter is high and sources current for the segment LED. When the driver transistor is cut off, the input of the inverter is high and the output low. The segment LED is darkened.

To use a common-anode display with a common-cathode driver IC, the circuit shown at D can be employed. When base current turns a transistor on, the voltage appearing across R causes the inverter output to go low and sink current through the LED. When the transistor is cut off, the input of the inverter is low and its output high. No voltage drop exists across the LED, no current flows through it, and it remains darkened.

If you use either of these converter circuits, be sure that the inverters can handle the amount of current required by the LED segments.

Interestingly, some manufacturers are now producing IC’s that will work with either common-cathode or common-anode displays. At E, National Semiconductor’s MM5402 and MM5405 clock chips’ output stage is shown connected to a common-anode display. The SEGMENT pin, internally connected to the drain of the driver, is wired to the cathode of a segment LED. The OUTPUT COMMON pin, internally connected to the source of the driver, is grounded.

At F, the common cathode application is shown. The OUTPUT COMMON pin is connected to the positive supply voltage and the SEGMENT pin connected to the anode of one segment diode in the common-cathode display. The diode cathodes are grounded. Thus, the driver either sinks or sources display current, depending on how the output pins are connected, making the clock chips compatible with either type of display. Another nice feature of these clock chips is their ability to be directly connected to LED displays. Current limiting resistors are not required.
Sabtronics Model 8100
Frequency Counter

100-MHz, 8-digit counter kit
sells for only $90

THE Sabtronics Model 8100 is one of
a number of frequency-counter kits
costing under $100 that have recently
appeared on the market. It has a guar-
anteed measuring range of from 20 Hz
to 100 MHz (to 600 MHz with an optional
prescaler, add-in facilities for which are
provided in the basic instrument). The
worst-case input sensitivity up to 100
MHz is specified at 25 mV rms. Three
selectable gate times permit the fre-
cency to be directly read from the in-
strument’s eight-digit numeric display.

Built into the counter is a crystal time
base that provides a rated measurement
accuracy of better than 2 ppm
(0.0002%), with a rated stability of ±1
ppm. The input impedance is 50 ohms
or 1 meegohm, switch selectable. Input
protection is to 150 volts rms up to 10
kHz, 90 volts rms between 10 kHz and 2
MHz, and 30 volts rms between 2 MHz
and 100 MHz. Also included is a three-
decade switchable attenuator for the in-
put, a nicety rarely seen in an economy-
priced frequency counter.

The Model 8100 frequency counter
measures 8” W x 6½” D x 3” H (20.3 x
16.5 x 7.6 cm). Supplied without batter-
ies (8 AA cells are required), ac adapter,
or input cable, it is priced at $89.95.
Available as options are an ac adapter
and a 600-MHz prescaler.

General Information. The Model
8100 is a simple frequency counter to
use and interpret. You simply set the
input attenuator switch to X1, X10, or
X100, depending on the level of the in-
put signal. Then you connect the input
cable to the desired points in the circuit
under test, select the appropriate gate
time, and read the frequency directly
from the display. Frequencies are dis-
played in kilohertz to the left and hertz to
the right of the decimal point. (The deci-
mal point automatically appears in the
proper location when the GATE TIME
switch is set to any position.)

To operate the instrument, you set
each of four switches to their appro-
priate positions. Power is applied to the
counter and input impedance is simul-
taneously selected by placing the switch

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labelled 50Ω/1MΩ/OFF to one of its first two positions. Then, depending on the frequency being counted, the DIRECT/ PRESCR. switch is set to one or the other position. (We built and tested the basic 100-MHz version of the counter. Hence, we always set this switch to the DIRECT position during our tests.)

The GATE TIME switch must be set next to the appropriate position according to the frequency being counted. (The GAT. LED in the display blinks on briefly once every tenth of a second, every second, and every 10 seconds for the 0.1S, 1S, and 10S positions of the switch, respectively.) At this point, the frequency being counted appears in the display. If an overflow condition exists, the OFL. LED comes on, indicating that the GATE TIME switch must be set for a shorter interval than that selected.

About the Kit. This was a very easy kit to assemble, requiring approximately four hours to complete from the time we opened the carton in which it was shipped until the end of calibration and final assembly. Credit for this simplicity goes to the open and uncluttered layout of all parts and the explicitly illustrated and well-written assembly instructions. Moreover, we encountered no difficulties in assembling the kit.

With the exception of the input connector(s) and the two bATTERY holders, every electronic component that makes up the counter is mounted on either of two silk-screened printed circuit boards. The large main pc board is double-sided and has plated-through holes. Most of the components go on this large mother board, which contains all counting and decoding circuits and the time-base oscillator. The smaller display board is single-sided and contains the seven-segment LED displays, the segment and digit drivers, and the operating switches. The only items needed for successful completion of the kit are a pair of long-nose pliers, diagonal cutters, a screwdriver, solder, and epoxy cement.

During assembly, we noted that all resistors had a 5% tolerance, even in those circuits where a 10% or 20% tolerance would have sufficed. Another plus is that the kit is supplied with sockets for all ICs, which reduces the possibility of damage to the ICs during installation and soldering.

Laboratory Tests. The frequency counter performed well within its published ratings in every important area of operation. Our test setup consisted of both a very-low-frequency and a 500-MHz r-f signal generator for providing the wide range of frequencies required for making a frequency-range test on the instrument. We also used a very accurate laboratory-grade frequency counter to check the Model 8100’s accuracy and a digital multimeter to monitor the signal level required for positive triggering.

Our frequency-range test revealed that the Model 8100 can accurately operate down to 1 Hz or less and to well beyond Sabtronics’ guaranteed 100-MHz high end. In fact, the high-frequency limit was 134 MHz before we had to boost the input signal level to an amplitude to 25 mV, which is the level specified by Sabtronics for a 100-MHz display. (The frequency counter actually operated unambiguously out to 148 MHz with a 50-mV input and might reasonably have responded well beyond 150 MHz if we had boosted the input signal level to 100 mV.)

Sabtronics specifies the input sensitivity of the frequency counter to be 15 mV rms from 20 Hz to 70 MHz and 25 mV rms from 70 to 100 MHz. In our tests, however, the actual sensitivity was much better. It measured 10 mV rms or less between 1 Hz and 75 MHz and was typically an average of 3.8 to 7.1 mV over most of this range. Between 75 and 120 MHz, it was between 7.5 and 17 mV. The sensitivity dropped to 25 mV at 134 MHz and 50 mV at 148 MHz.

We performed the above tests first with the no-instruments calibration procedure detailed in the assembly manual supplied with the kit and again after precision calibration with instruments. When we compared the two sets of test data obtained, there was no significant difference in performance. (The test results given above were for calibration without instruments.)

We did not check the average aging rate of the instrument, which is rated at ±1 ppm/month for the first three months of continuous operation and ±5 ppm/month thereafter.

User Comment. Having used the Model 8100 frequency counter for several months, both on our workbench and in the field, we have come to appreciate its reliable and accurate performance under just about any testing condition imaginable. It is lightweight, portable, and relatively rugged. There is no doubt that the Model 8100 is suitable for experimenters, hams, and service personnel in shop or field.

Circle No. 104 on free information card
"THE REGULAR audience for shortwave broadcasting is vanishingly small." That’s the position of the ham radio lobby, ARRL, expressed in its Oct. 1978 issue of the magazine QST. In an effort to bolster its own threatened position at the 1979 World Administrative Radio Conference, ARRL has tried to show that no more frequencies should be allocated for international shortwave broadcasting.

The ARRL position is based on a report it commissioned from the Stanford Research Institute, and then submitted to the FCC. In the Review of International Broadcasting, Lawrence E. Magne effectively debunks the S.R.I. report, showing among other things that the figures it gives on the size of the U.S. SWL audience are only a fraction of those obtained in a 1975 Gallup Poll commissioned by Radio Canada International. The previously confidential Gallup Poll figures were first published in Magne’s article.

The poll showed the following total figures for the U.S. adult audience of six leading international broadcasting stations. They are composed of three subtotals—those who listen less than once a month; those who listen one, two or three times a month; and those who listen once a week or more often. (The adult population of the U.S. was taken to be 142,000,000 as of 1975.)

- BBC: 4,544,000
- Radio Canada International: 4,402,000
- Radio Moscow: 2,840,000
- Deutsche Welle: 2,414,000
- Radio Habana Cuba: 2,272,000
- Radio Nederland: 1,988,000

These are not huge numbers, but they are far from being "vanishingly small" and only cover the audiences for six out of hundreds of stations. Moreover, the U.S. population is larger today, and SWL is enjoying a surge in popularity. Combined with many new easy-to-tune shortwave models, the audience can only be larger today. Does anyone think that Radio Shack, Panasonic, Sony, and Lafayette would introduce new SW models, as they recently have done, if the market were shrinking?

Future Plans. For further evidence of the viability of international shortwave broadcasting, we need only look into these "future plans," all of which at press time had not yet taken place. Any dates mentioned should be considered approximate since construction invariably takes longer than the station had originally expected.

- Adventist World Radio is getting into shortwave broadcasting in a big way after several years of buying time on existing stations. AWR plans to build its own shortwave station in Liberia. Construction of four transmitters totaling 500 kW to begin sometime this year. Already being built is another AWR station in Guatemala on AM and FM, with a separate shortwave service to be on the air in the 25-meter band by mid-1979.
  - Trans World Radio, Monte Carlo, is replacing a pair of 100-kW shortwave transmitters with 500-kW units.
  - HCJB, Ecuador, is installing a new 10-kW tropical band transmitter to provide regional service in Spanish and Quecha, probably on 3220 kHz, targeted for May. But HCJB minimizes its involvement in the Hawaiian shortwave project mentioned in this column last November.
  - It had been reported that Radio Voice of the Gospel, which was nationalized out of Ethiopia, would make a comeback from Gabon’s new set of 500-kW shortwave transmitters. However, those senders have been slow in coming into regular service and RVOG is keeping its options open.
  - Ever-eager to make missionary inroads into China, while avoiding the risk of nationalization, is the Far Eastern Broadcasting Co., which is installing two 100-kW shortwave transmitters on the island of Saipan, which FEBC emphasizes is "American soil."
  - Radio-TV Dominicana is gearing up for an expanded international service, probably initially on 5970 and 9505 kHz, with programs produced by an Avery Schreiber look-alike, Teo Veras. The new service is to include DX features both from local and U.S. sources.
  - Radio Mexico has expanded its transmission hours, showing up on new frequencies (including two already occupied by AFRTS), and has announced plans to begin broadcasts in English and Japanese. Another shortwave station is expected at the Autonomous University of Nuevo Leon, in Monterrey, probably on the 49-meter band.
  - Radio Nacional, Venezuela, may upgrade its irregular international broadcasts from 10 to 50 kW on 15400 kHz. But this is eclipsed by the acquisition of a million-watt medium-wave transmitter to be installed on the Paraguana Peninsula and used for international broadcasts in several languages. If its frequency is clear, this should have no trouble getting into the USA. The transmitter operated formerly for a short time on 625 kHz from Costa Rica.
  - Radio Free Europe has kept in service some 10-kW transmitters at Holzkirchen, West Germany. They’re no match for the megawatts of jamming they may face, so plans are to replace these with 250-kW senders.
  - Something is going on in Sri Lanka. It had been widely reported that West Germany’s Deutsche Welle was building a

MAY 1979

By Glenn Hauser
shortwave relay station here, perhaps to be used in conjunction with other stations, such as the Voice of America (which already has a low-powered station in Colombo), and Radio France International (which is also building a relay in French Guiana). But a DW spokesman says this is merely speculation, although DW does plan to put on a relay station somewhere in Asia.

- Mongolia is expanding its domestic radio network, with Soviet aid. New stations have already gone on the air at Altay and Choybalsan, on 4995 and 4850 kHz, and more are expected.
- BBC plans to set up a new relay station in Lesotho, to serve southern Africa.
- Radio Nederland is upgrading its domestic site near Lopik with several new 500-kW transmitters.
- Radio Australia has installed two new 100 kW Harris-Gates transmitters at its main site in Shepparton, Victoria, allowing older transmitters to be overhauled a pair at a time, with an ultimate gain of two senders when all are back on the air. And the Darwin relay station is to be revived, after several years of silence following cyclone damage.
- Upgraded international services are also in the offing from many other countries, including Kenya, New Zealand, Norway, Saudi Arabia, Sweden and Yugoslavia.
- Not all the news is good. Radio Canada International has suffered a $1.5 million budget cut, part of a $71 million overall CBC budget cut. RCI has had to cut its staff by 15%; that is, 36 people declared redundant; and RCI is obliged to move out of its quarters in the Maison de Radio-Canada in the city of Montreal.

**Conventions.** The big event each year for shortwave listeners is the annual ANARC (Association of North American Radio Clubs) convention. Anyone who is interested in DX listening is welcome to attend. No club membership is required. This year's convention takes place in Minneapolis, June 22-24. Equipment representatives are expected as well as personalities such as Bob Zanotti from Swiss Radio International, Ian McFarland of RCI, Stewart Spencer of the VOA, and Alfonso Montalegre of Radio Nederland. For full details, send a legal-size SASE to: ANARC 79, 3320 Grand Ave. South, No. 305, Minneapolis, MN 55408.

The equivalent in Europe takes place June 1-4 in Vienna. For details, contact EDXC Conference Committee, P.O. Box 11, A-1111 Vienna, Austria. Specially low accommodation rates are promised.

**Updating Listings.** The following changes and additions should be made in the "English Broadcasts" listings in the April issue:

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<th>Station</th>
<th>Frequencies, changes</th>
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<td>R. Australia</td>
<td>Not 0800-, also not 9540</td>
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<td>1100-1156</td>
<td>R. RSA</td>
<td>Not Sun. and not 21535</td>
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<tr>
<td>1100-1245</td>
<td>TWR, Bonaire</td>
<td>15255, ex-11815 (Sat.-1330, Sun.-1415)</td>
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<tr>
<td>1100-1330</td>
<td>BBC</td>
<td>9510, ex-5990</td>
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<td>1230-1550</td>
<td>WYFR</td>
<td>21525, 17845 (Sun. only)</td>
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<tr>
<td>1400-1430</td>
<td>V. Rev. Party</td>
<td>4109, ex-4120</td>
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<tr>
<td>1500-1730</td>
<td>R. Australia</td>
<td>11870</td>
</tr>
</tbody>
</table>

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<thead>
<tr>
<th>GMT</th>
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<tr>
<td>1200-1230</td>
<td>Israel Radio</td>
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<td>1700-1730</td>
<td>R. Pakistan</td>
<td>11672, 9465</td>
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<td>1700-1800</td>
<td>WYFR</td>
<td>21615, 17845, 15160</td>
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<td>BRT Belgium</td>
<td>17745, ex-17735</td>
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<td>1800-1900</td>
<td>WYFR</td>
<td>21615, 21525, 17845, 15130</td>
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<td>1800-2100</td>
<td>R. Kuwait</td>
<td>15345, ex-12085 (freq. changes)</td>
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<td>1800-2130</td>
<td>AFRTS</td>
<td>21570</td>
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<td>1900-1930</td>
<td>R. Afghanistan</td>
<td>11985, 11890, 15140, 15295</td>
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<td>(freq. changes)</td>
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<td>1900-2100</td>
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<td>2000-0230</td>
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<td>15070</td>
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<td>R. Can. Int.</td>
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<td>2115-0030</td>
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<td>VOA</td>
<td>26095, not 26895</td>
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<td>0000-0100</td>
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<td>R. Kiev</td>
<td>15240, 15100, 11790, 9580; not 15180, 12000, 7150, 6020, 5980</td>
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<tr>
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<td>BBC</td>
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<td>0145-0215</td>
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<td>R. Lebanon</td>
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<td>WYFR</td>
<td>5980</td>
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<tr>
<td>0300-0600</td>
<td>V. of Voyageur</td>
<td>Probably off the air</td>
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<td>0400-0700</td>
<td>AFRTS</td>
<td>15430, 15330, 9685; not 11805, 11790</td>
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<td>0430-0500</td>
<td>R. Sofia</td>
<td>9765, not 9530</td>
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<tr>
<td>0500-0700</td>
<td>HCJB</td>
<td>9745, ex-9560</td>
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<td>0600-0700</td>
<td>R. RSA</td>
<td>Add 21535</td>
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<td>0615-0630</td>
<td>R. Can. Int.</td>
<td>11775, 9590; not 11845, 9635</td>
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<td>0645-0700</td>
<td>R. Can. Int.</td>
<td>11775, 9590; not 11845, 9635</td>
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<tr>
<td>0700-0800</td>
<td>Malta Calling</td>
<td>9670 (Sat. only)</td>
</tr>
</tbody>
</table>

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**Computer Bits**

By Hal Chamberlin

**16-BIT MICROPROCESSORS**

RECENTLY, there has been a great furor over the announcement of new 16-bit microprocessor ICs. The truth is that there are many 16-bit microprocessors already available so why should manufacturers, users, and the press take such a sudden interest in more of the same—if, in fact, they are the same? To find the answer to this question, let us briefly describe the 16-bit processors—past, present, and upcoming—known to this author. We will then see if such a summary, though it can't deal with each unit in depth, will show real differences among them.

**IMP-16 and INS8900 (PACE).** The IMP-16 was the first 16-bit microprocessor introduced (1974) and is still one of the most powerful in wide use. Its instruction set resembles that of the Data General NOVA minicomputer, but has many enhancements such as memory-to-register arithmetic. Its speed is 7 µs for a memory-to-register add instruction. Full 16-by-16-bit hardware multiply is available as an option and executes in about 150 µs. Memory up to 64K words (128K bytes) can be addressed.

Supplementing the IMP-16 in new designs is the INS-8900, which is a single IC rather than five or six devices. The instruction set is essentially the same as the IMP, but it has provisions for unlimited stack depth and five different interrupts. Hardware multiply/divide is not available. The standard version is slower than the IMP (10 µs add rather than 7 µs) though a twice-as-fast version has been rumored. In 1976, there was an abortive attempt to design a modular hobbyist oriented system around the PACE microprocessor.

**MCP1600.** This is a three (or more) IC set that may be *microprogrammed* by the manufacturer to mimic nearly any 16-bit processor desired. First available in 1975, this chip set is the basis of the LSI-11 microcomputer manufactured by Digital Equipment Corp., which is used by Heath in its H11 microcomputer. The instruction set is precisely that of the PDP-11/40, a very powerful minicomputer. The LSI-11 can perform a 16-bit add from memory in 5.6 µs. Hardware multiply (60 µs), divide, and floating point are available as an option. Maximum memory size is 56K bytes excluding those addresses assigned to I/O.

Western Digital, creators of the MCP1600, also has microprogrammed the set to emulate a NOVA minicomputer. Although instruction timing is not available, it should be in the 4 µs range for an add because of the much simpler NOVA architecture. Alpha Microsystems has also written a microprogram for its proprietary instruction set. Its AM-100 16-bit microcomputer has been on the market for more than a year and uses the S-100 bus. The AM-100 is claimed to be significantly faster and more powerful than the LSI-11.

**CP-1600.** A single-chip 16-bit microprocessor, the CP-1600 has been around nearly as long as those just described. Its architecture and instruction set resemble those of the PDP-11, but is substantially simplified to keep costs down. The standard version performs an add from memory in 4 µs, while a reduced cost version ($8 each in 100's makes it the least expensive 16-bitter so far) takes twice as long. One feature that has probably done the most to discourage popularity is its use of a 10-bit instruction word. This leads to inefficiencies in general-purpose systems where programs are stored in 16-bit read/write memory. Up to 64K words of memory can be addressed, however.

**MICRO NOVA.** This one-chip microprocessor, now in its third year, uses the NOVA-3 instruction set precisely. The manufacturer is Data General, developers of the NOVA minicomputer in the late 1960's and one of only three mini-

**POPULAR ELECTRONICS**
computer manufacturers who also make microprocessor ICs. Since the NOVA instruction set does not include an add-memory instruction, its speed is difficult to compare. Load from memory, however, takes 2.9 µs while register-to-register add requires 2.4 µs. Hardware multiply is accomplished in 42 µs. The NOVA instruction set normally provides for addressing of 32K words of memory, although a special mode that inhibits multilevel indirect addressing allows 64K. Many designers have avoided the Micro Nova because of its odd non-TTL logic levels and the four supply voltages (+14, +10, +5, -4.25) required.

9900. The 9900 is probably the most popular of the currently available 16-bit microprocessors. Made by Texas Instruments, it is simultaneously innovative, powerful, and easy to use. The full 16-bit version boasts an instruction set similar to the PDP-11, but with an increase to 16 general-purpose registers and a unique memory-to-memory architecture. The registers are actually kept in memory, which allows very rapid response to interrupts by simply shifting the portion of memory devoted to registers. Since the number of on-chip registers is drastically reduced, a smaller, more economical IC chip is the result. Even though an "add from memory" requires no fewer than five memory cycles, it is executed in a respectable 6 µs. Hardware multiply/divide is standard and multiply time is a very speedy 18 µs. The input/output mechanism uses a unique semiserial technique that obviates the need for special I/O chips to accomplish simple I/O ports. Another striking feature is the 64-leap package that allows full 16-bit address and data buses without multiplexing. Because of its byte addressing feature, only 32K words of memory can be addressed.

The 9900 has been available to hobbyists for more than two years in the Technico System 16. The forthcoming TI personal computer is rumored to utilize the 9900 for its CPU.

9440 Microflame. This is another, though considerably newer, NOVA emul- lator microprocessor. Compared with the Micro Nova it is faster (2.4-µs load and 1.25-µs add) and much easier to interface and power. Hardware multiply/divide is not available in the 9440. It is unique in that bipolar integrated injection logic (BIL) is used rather than p- or n-channel MOS logic. A model (9445)

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that will be three times as fast is planned. The name "Microframe" has given rise to all sorts of whimsical names for associated products. Examples are the "Firebug" debugging system and "Spark 16" computer.

6809. At this point we start getting into microprocessors that have been announced but are not yet generally available. The Motorola 6809, for example, is an enhancement of the popular 6800 8-bit microprocessor. Its status as a 16-bit microprocessor is arguable, however, since the data bus is only 8 bits wide. This means that the efficiency advantage of 16-bit instructions is only partially realized, although it does make retrofitting to existing hardware easier. The 6809 retains all of the original 6800 instructions. Operation codes have been changed, however, which necessitates reassembly. Improvements include the addition of a second index register, another stack pointer, and a relocatable zero page. A 16-bit add from memory requires 5 \( \mu \) s with the standard 2-MHz clock frequency. Hardware multiply is included. It is only 8x8, but the 10-\( \mu \)s speed allows 16x16 multiply at a speed comparable to earlier 16-bit microprocessors. Memory up to 64K bytes can be addressed.

8086. This is Intel's entry into the current 16-bit microprocessor race. Although source code compatibility with its 8080 microprocessor is claimed, it is only through a rather complex translating assembler. Besides some carry-overs from the 8080, the instruction set is unique and as powerful as those of current minicomputers. Compared to the 8080, the biggest improvement is the inclusion of numerous addressing modes, though relative and indirect through memory are not available. The average speed of the 8086 is impressive: 1.6 \( \mu \) s for a memory-to-register add and a mere 375 ns for a register-to-register add. Note the use of the term "average"; part of the speed improvement is owed to an instruction "lookahead" circuit which is rendered ineffective when a lot of conditional branch instructions are executed.

A definite departure from what we have seen so far is the ability to address 1-million bytes of memory! This extended addressing capability is through a memory bank switching scheme, however. Hence, only 128K (64K program storage and 64K data storage) can be reached by a program without the hassle of using a bank switch.
Z8000. The Z8000 is Zilog's 16-bit contender. The Z8000 instruction set gives nearly every possible combination of instruction type and data lengths of 4, 8, 16, and 32 bits! Yes, there are instructions that deal directly with 32-bit operands and registers. Naturally, with such instruction-set sophistication, hardware multiply and divide are available with up to 32-bit operands as well. Sales literature compares Z8000 speed with the PDP-11/45 minicomputer (a popular but expensive minicomputer that fills a rack) and declares the Z8000 winner with an add time of 1.75 µs. Multiply is less speedy in comparison (17.5 µs for 16x16 and 88 µs for 32x32) but is still quite respectable for a single-chip microcomputer. Up to 8-million words of memory can be directly addressed by the Z8000. This is made possible by the 32-bit registers, which is much more convenient than bank switching.

MC68000. This last processor is also the most powerful and farthest from being available. In reality it is a 32-bit machine with a 16-bit data bus. All 8 accumulators and 8 index registers are a full 32-bits in length. This puts it in the maxicomputer league along with the IBM 370. At this time, exact specifications are not available, but the add time is stated to be 1.5 µs. Multiply/divide is said to be faster than the Z8000, but no figures are available. Programs can directly address up to 16-million bytes of memory through use of the 32-bit index registers and a 24-bit program counter.

Conclusions. By now it should be obvious why there's a great interest in the recently announced 8086, Z8000, and MC68000 microprocessors. These machines are at least three times faster than existing 16-bit units. Their ability to address vast quantities of memory promises to once more fill up computer cabinets with memory, this time with 64K rather than 4K boards. The latter property makes programmers happy; the former makes everyone happy.

In short, the new microprocessors give more of what 16-bit (and 32-bit) architecture is good for. But don't expect to find a system using the new chips in computer stores right now. Of the top three, only the 8086 has actually been manufactured, so it might be well into 1980 before personal systems using these chips are available. Meantime, LSI-11 and 9900-based systems still greatly outperform 8-bit-based systems and are available now.
**WordWizard.** This advanced text editor, designed for use with this firm's Helios disk system, has a rather large number of features for document activity (create, edit, print, merge, etc.), editing functions (cursor control, tab, justify, search and block moves, etc.), and format statements (spacing, line formats, pagination, etc.), far too many to mention here (brochure available). Up to 110 text pages can be stored on a diskette, and horizontal scrolling allows viewing lines up to 128 characters long. All documents may include comments which will not appear on the printout. $295 with manual and two diskettes (Helios). Processor Technology, 7100 Johnson Industrial Drive, Pleasanton, CA 94566 (Tel: 415-829-2800).

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The LM3909 was originally designed as an LED flasher, but has many other applications. One that I've enjoyed experimenting with is a miniature power supply that allows a tiny watch battery to power a neon lamp or even a powerful semiconductor-laser pulser. Both these applications require 70 to 150 volts at relatively low current.

Figure A shows the circuit of the LM3909 dc-dc converter. In operation, the LM3909 rapidly switches Q1 on and off at a rate determined by C1. The transistor can be considered a switch in series with choke L1 and resistor R2. Each time Q1 switches off, the magnetic field set up by the current flowing through L1 collapses and induces a high voltage across the inductor. This voltage is rectified and stored in C2.

The LED is a bonus feature of the circuit. It glows to indicate when the circuit is operating. The neon lamp and 15,000-ohm series resistor shown in Figure A are optional. They provide a visual indication that the circuit is producing 70 or more volts. When powered by a 1.2-volt nickel-cadmium or 1.35-volt mercury "button" cell, the circuit produces enough voltage to flash the lamp when it is connected across capacitor C2.

If you don't like the orange glow of a neon lamp, try a green neon lamp (Radio Shack 272-1106 or equivalent). This lamp has a phosphor coating on its inside surface that glows green when illuminated by the radiation produced inside the lamp. In any case, be sure to use a quality lamp because some of the surplus neon lamps I've tried do not work well.

The key components of the circuit are L1 and R2. In the prototype circuit, I used a miniature Essex choke with an inductance of 1000-μH for L1. This choke is about the size of a 1/4-watt carbon composition resistor. If this choke is used, the resistance of R2 should be between 75 and 85 ohms.

If you can't find this choke, experiment with others until you find one that produces enough voltage to light a neon lamp. You'll find that many different chokes will produce a useful output. One version of the circuit that I built uses a miniature 33-mH choke (Aladdin) with excellent results. The 1979 Allied catalog (401 E. 8th St., Fort Worth, TX 76102) lists a number of subminiature r-f inductors on page 145 that should work fine.

If you don't use the 1000-μH choke specified, you'll need to experiment with the value of R2. As the inductance is increased, R2's value can be decreased. Actually, R2 is not even necessary beyond a few millihenries.

Assembly of this circuit should present no problems once you've selected a choke and determined the resistance of R2 (if it is necessary). I used a piece of perforated board with copper solder pads at each hole. Figure B is a pictorial view of the assembled circuit.

Begin by inserting the components into the top side of the board and interconnecting their leads with wrapping wire. Then solder all the connections to their respective solder pads. Figure C is a photograph of the complete prototype. This circuit includes a neon lamp and series resistor to illustrate its operation as a dc-dc converter. Don't forget that the circuit has many other possibilities.

For example, most semiconductor lasers require current pulses of many amperes for proper operation. The circuit in Figure A can power a four-layer-diode laser pulser with ease, especially if L1 has an inductance of 10 to 35 mH. Recently, I built a midgap laser transmitter using the circuit in Fig. A as a power supply (L1 = 33 mH, no R2). The circuit is completely self-contained and includes lens, mercury "button" cell and switch in a 0.5" x 3" (1.3 cm x 7.6 cm) brass tube. If there is sufficient interest, I'll describe its construction as a future Project of the Month.
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<td>$301-$1000</td>
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Transistor Checker

- Completely Assembled
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- The ABI Transistor Checker is ideal for diagnosing and identifying transistors. It can help you identify the base and collector terminals of a transistor by connecting it to the base and collector leads of the transistor.
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- Operating costs are $19.95.
- The panel is an 8-bit system.

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- Lets you work with both hands.
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- Clamp "3rd Hand" on edge of bench, table or workbench, hub or circuit board, or any component.
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J7E01

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- With light switch to allow you to see the time in the dark.

$65.50 ONE YEAR FULL WARRANTY

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Assembled not a kit

Features:
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- 12 hours real time format
- 24 hours alarm audio output fail-safe
- Power failure indicator
- Count down timer 59 mins
- 12/6V AC 50/60 Hz input
- 10 min snooze control

Price: $8.50 EACH

Transformer: $1.75

NEW MARK III 9 Stage Color LED TV

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IN KIT FORM $18.50

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KIT FORM $8.75 EA.

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KIT FORM $7.95

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power transformer: $9.50 ea.

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- Power interrupt indication
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- 110V AC 60Hz input
- Factory assembled. NOT A KIT $17.50 EACH

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- K's lead controlled circuits
- DC powered (15V battery)
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- 24 hr. alarm set
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- Dual time zone display
- Stop watch function

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- Featured in Oct. Popular Electronics
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<th>Design</th>
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<td>Econoram Xl-16X</td>
<td>8K X 8</td>
<td>S-100</td>
<td>static</td>
<td>2 MHz, 4 MHz, 2K boundaries</td>
<td>$369</td>
<td>$419</td>
<td>$469</td>
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<td>16K boundaries</td>
<td>static</td>
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<td>$699</td>
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<td>S-100</td>
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<td>2 MHz, 4 MHz, 2K boundaries</td>
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<td>$164</td>
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<td>16K X 8</td>
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<td>12K X 8</td>
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<td>4 MHz, 4-8K, 4-16K</td>
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