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Computer Review
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COMPONENT NOTATION AND UNITS
We normally specify components using an international standard. Many readers will be unfamiliar with this but it’s simple, less likely to lead to error and will be widely used everywhere sooner or later. ETI has opted for sooner!

Firstly decimal points are dropped and substituted with the multiplier; thus 4.7uF is written 4.7. Capacitors also use the multiplier nano for nanofarad (1 nano farad = 1000pF). Thus 0.1uF is 100nF, 5600pF is 0.56uF. Other examples are 5p6 = 5n6 and 0.5pF = 0.5p. Resistors are treated similarly: 1.8Mohms is 1M8, 5.6kohms is 5K6, 100ohms is 100k. Please note we do not keep track of what is available from who so please don’t contact us for information on PCBs or kits. Similarly do not ask PCB suppliers for help with projects.

PCB Suppliers
ETI magazine does NOT supply PCBs or kits but we do issue manufacturing permits for companies to manufacture boards and kits to our designs. Contact the following companies when ordering boards.

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ETI-MARCH-1984-5

POSTAL INFORMATION
Second Class Mail Registration No.3955. Mailing address for subscription orders, undeliverable copies and change of address notice is: Electronics Today International, Suite 501, 525 Queen Blvd., Toronto, Ontario, M4H 1B1.
An intriguing press release has arrived from Kodak Canada Ltd. Under a licensing agreement from Drivetece Inc. of California, Kodak of New York will begin manufacturing a 5.25 inch "Superminiflop-py" disk drive system. The drives will be double-sided with 192 tracks per inch, and a capacity of — wait for it — 3.33 megabytes. It can also read 48 and 96 track-per-inch disks. They'll be available in the first quarter of 1984, and are offered for the initial price of $350 (U.S.) to computer enthusiasts. Do you live life in the fast lane, and can't be far from a computer? Radio Shack announces the portable version of their Model 4, The Model 4P. It features a 9-inch screen, and can run all Model III TRS-80 and LDOS programs via 5.25 inch drives. Also offered is memory expansion to 128K, CP/M, and a modem. The keyboard passes across the front to protect the CRT and drives. At your local Radio Shack computer dealer.

1984 IC MASTER

The 1984 edition of the IC MASTER has been published by Hearst Business Communications and is available in Canada from Future Electronics. The newly revised deluxe 2-volume edition contains over 4,000 pages and lists key specifications for 35,000 integrated circuits, microcomputer boards and related products made by 225 manufacturers. Both new and discontinued devices are shown in the alternate source directory. This section provides information on replacements and lists approximately 55,000 IC substitutes. More than 60 manufacturers have supplemented the editorial material and tables in the IC MASTER with extensive data sheet sections. The eleven technical data sections of the 1984 IC MASTER, organized by function and key parameters are: Military, Digital, Interface, Linear, Memory, Microprocessor, Microcomputer Board, Microcomputer Support Board, Microcomputer Development System, Custom/Semicustom IC's and PROM Programmers.

According to the Toronto Globe and Mail, Coleco Industries Inc. of Hartford, Conn., says it plans a new attachment that would enable its Adam home computer to run some software designed for the Personal Computer of International Business Machines Corp of Armonk, N.Y. Coleco said it would introduce the plug-in attachment late this year. It would not say how much software designed for the IBM PC would be able to run on the Adam with the new attachment.

AN ALTERNATE source to the OP-27, the industry's lowest noise high-precision operational amplifier, has been introduced by Analog Devices, Inc. The AD OP-27 combines maximum input noise of 0.18uV peak-to-peak with maximum offset voltage and offset voltage drift of 25uV and 0.6uV/°C, respectively.
Compact Disk Player
Sony has introduced a new compact audio disc. The CDP-200 has many of the same features as the CDP-101, as well as a new index search feature to help classical music buffs zero in on specific passages within individual movements. All key component parts — including the critical digital-to-analog converter LSI and the semi-conductor laser itself — have been developed by Sony. The CDP-200 features horizontal front loading, utilizing Sony's Linear Skate Drawer mechanism. Feather-touch controls on the front panel are ergonomically designed, and include Play, Pause and Reset/Stop. In addition, Automatic Music Sensing (AMS) allows instant access to any adja-

cent musical selections. The Fast Scan mode enables the user to find specific portions of a selection quickly by playing a constant stream of musical "samples" as the laser sweeps across the disk. The Sony CDP-200 also has a special Index search feature to help the user locate a specific subcoded portion of a long classical movement. Other features in-
clude: a two-way repeat function that can be programmed to replay either an entire selection or any amount of music located between an A and B point; a concentrated display that provides the track and index number, plus both elapsed time and remaining information; automatic disk sensing protection; and headphone output jack with level control. GTE said today that it will test Sylvania Metalure® lamps in space during the next Space Technology Shuttle mission scheduled for January 30 to February 7, 1984. "The purpose of the space experiment, which is part of NASA's 'Get Away Special' program, is to observe the operation of the efficient Sylvania lamps in a zero gravity environment," said James F. Cosgrove, vice president of research and engineering at GTE Lighting Products. "Data collected on the behavior of the lamp arc without the disturbances of gravitational convection forces should help us in design of even more efficient lamps," he stated.

Missing PCBs.

Our apologies for the omission of the ZX81 ADC and the Digger PCBs from our January issue. The person responsible has been flogged and put on short ra-
tions.

An effective telecommunications system is vital to the growth and prosperity of any nation, Basil A. Benetoeu, vice-chairman of the board of Northern Telecom Limited, told the plenary session of the Pacific Telecommunications Conference held in Honolulu recently. Discussing the impact of telecommunications on the economies of developing Pacific Rim nations, Mr. Benetoeu said, "Nations with only minimal telecommunications tools are inhibiting their prospects for productivity improvement, industrial development, and competitiveness in an increasingly competitive world."

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IN TYPING up an important contract, a stenographer realizes that the client didn’t specify whether he wanted a thousand model-156A or a thousand model-156B widgets. And the sales rep who closed the deal is on the road.

No problem. She phones the sales rep as he speeds along a highway somewhere. Without slowing down, he discusses it with her and decides to ring up the client himself. The client takes the call as he’s inspecting the unfinished fourth floor of a building under construction. He checks in turn with his own head office in another city and phones the information back to the travelling sales rep, who relays the order to the waiting steno. All done without searching for telephone booths or disrupting anyone’s schedule.

This is how cellular radio would work from the users’ perspective — as a means of calling between any locations regardless of whether phone lines are handy.

The handheld receiver even looks like a telephone, a cordless model with a short antenna. It fits in a briefcase, on the dashboard or in a large pocket. To operate cellular radio, the caller punches in a number as on any push-button phone and presses another button to speak.

However, the technology to handle the call is quite different from normal telephone communications. As our speeding sales rep is straightening out his client’s order, a complicated process is carried out behind the scenes to keep the parties in contact.

Radio signals are transmitted to, or received from, the car phone by an antenna at a base located in the car’s current vicinity. When the mobile unit communicates with the stationary office phone, the signals are converted at the base and passed through the usual telephone lines. When it’s in touch with another remote unit, such as at the construction site, the signal is relayed to a base in the receiver’s vicinity and broadcast to him from there.

As the car travels out of range of the antenna, a computer automatically reroutes the call through another base whose area the car is entering. As the sales rep drives around, his calls are repeatedly switched from base to base without noticeable interruption. Moving from one urban centre to another is more difficult but possible.

Made feasible by the marriage of radio, telephone and computer technologies, the roaming capability of cellular radio allows communication between any parts of the country where a cellular
system is established. The regions are divided into honeycombs of cells, each cell up to 10 miles wide. Combined with a fourth technology, communications satellites, cellular radio can eventually be extended to remote areas that lack ground-based cellular systems.

What the DOC Ordered

Heralded as the 'next big thing after Pay TV,' cellular radio is coming to Canada. By the time you read this, the Department of Communications (DOC) might have already approved licences for setting up cellular radio systems across the country. The Alberta Government Telephone Company has been carrying out a test with 200 mobile phones in Edmonton since last December.

Some companies say they can have cellular radio systems up and running in major cities by 1985. Although initial costs will limit the appeal to businesses for a few years, falling prices may eventually attract a wider market. A DOC regional director predicted cellular equipment will eventually become consumer items sold through stores like Eaton's, Simpsons and Radio Shack.

Cellular radio has already been established in some countries. The first commercial system was set up in Sweden, Norway, Denmark and Finland and already has over 30,000 users. Japan has a cellular system running, and Australia has it under development.

The Scandinavian system illustrates that cellular radio can cross borders as easily as conventional telephone service, provided the systems are internationally compatible. Canada is establishing its system parallel with the United States where the Federal Communications Commission (FCC) is licensing metropolitan areas. The DOC is working with the FCC to create a continuum of frequency and cell structure across the two countries. Compatibility will facilitate cross-border communications and help Canadian cellular equipment manufacturers find markets in the U.S., Ottawa hopes. Judging by past experience, the reverse is also possible. American products could flood our market. The DOC has let applicants know they want to see Canadian-produced equipment worked into their plans.

The licencing process began last October with the DOC's call for applications to run "a high-capacity mobile radio system" in 23 urban regions across the country.

Two systems are to be licensed in each area — one operated by the local telephone company and one by a non-telephone company. When a call is placed from a mobile phone the pulse is detected by both systems in the area, but only one will respond, depending upon the frequency which identifies the system the caller subscribes to. Both systems, however, must interconnect with the existing telephone network.

By February, the DOC was flooded with applications, including submissions from alliances of some of the largest electronics and communications conglomerates in Canada, encouraged by projections of long-term multi million dollar profits. Within a month the possibilities were narrowed down to seven private applications plus the telephone companies. In August, communications minister Francis Fox announced that only a single national licence would be awarded the non-telcos.

The Soft Cell

Cellular radio has been called 'the rich man's mobile phone service.'

The first units will probably cost $2,500 to $3,500, or lease for $100 to $150 a month. This does not include airtime charges which could add an extra $150 per month to the bill.

The high price is said to reflect the massive expense of installing the cellular system. Figures in excess of $150 million have been quoted, not to mention equal amounts spent on research and development prior to installation or the continuing costs of maintaining the system.

Like other electronic products, cellular radio is expected to become cheaper in the years following introduction. With prices settling below $40 per month plus airtime charges, one percent of the population could be part of the cellular system by 1988, according to some predictions.

One licence applicant, CNCP Cellular Communications, owned by CNCP Telecommunications and Motorola Canada Cellular Inc., submitted a brief to the government that said such subscriber forecasts appear 'unduly optimistic' although they remained convinced that cellular telephones would benefit all Canadians.

Certainly the applications anticipate large profits can be made. Michael Kedar, a communications consultant with KVA Communications and Electronics Co., which was involved in the application by Cel Tel Corp., said, 'It's definitely a big potential money-maker. The telephone companies have monopolized this field but today cellular radio has opened it up.

Kedar sees it in the context of related technologies that consumers might become familiar with. At the lowest of three evolutionary levels are the cordless phones with mobility up to 1,000 feet from your home phone. These are coming into fashion now. The intermediate level, before cellular radio, is personal radio service with up to 15 miles range. This would be adequate for many users since studies have shown that 80 percent of people do not wander more than five miles from their home, according to Kedar.

Pre-cellular mobile phones are currently operated by about 50,000 Canadians through Bell Canada and the other telcos. Using mainly frequencies in the 400 MHz band, only a handful of these units can place calls simultaneously in a large city. A caller is as likely to get a busy signal as to have the call put through.

One of the DOC's stated goals with introducing cellular radio was to alleviate this shortage of frequencies for mobile telephone service. The 800 MHz band, where more frequencies are available, has been set aside for cellular use. (One side effect has been the removal of UHF televi-
Cellular Radio

The Remote Switch Group (RSG) co-ordinates the control switches. Illustration courtesy of ITT.

The above systems come with warranty and optional extended warranty.

Another advantage of the cellular method of dividing up the market, is that it is infinitely expandable. As the number of users reaches the limit in an area, the cell can be subdivided into two or more smaller cells with their own bases and channels.

ETI, Phone the Office

Mobile phone service is only one function of cellular radio, although it's likely to be the most widely used. Teleconferences and dozens of other enhanced services have been promised. Some futurists foresee a day when cellular phones will be as common as wristwatches and may in fact be worn on the wrist in Dick Tracy fashion. Everyone could get an individual telephone number that stays with them for life. Perhaps the ubiquitous SIN will double as the phone number.

Despite the benefits offered by cellular radio, one drawback for the harried sales rep comes to mind. When he can carry a phone in his pocket, it'll be harder than ever to get away from it all.

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Will the real Buck Rogers please sit down? Roger Allan looks at a technology which, while having promise, is nowhere near being a threat yet.

A PARTICLE BEAM is a stream of highly energetic atomic or subatomic size particles such as electrons, protons, hydrogen atoms or ions. By comparison, laser beams are composed of radiant energy photons. Presently, aside from potential applications as weapons, particle beam machines have potential for use in inertial confinement fusion for energy generation, nuclear weapons effects simulation, heating and welding, high intensity microwave generation, geophysical investigations, energy transmission, medical treatment, and basic physics experiments.

Consideration of the use of particle beams as weapons actually dates back to World War II, but the technology of the times could not support such a concept. The early U.S. Department of Defense efforts on particle beams were started by the Defense Advanced Research Projects Agency (DARPA) in 1958 under a program codenamed SEESAW.

The initial potential "mission" was ground-based strategic defense against ballistic missiles (ABM), with work centered at the Lawrence Livermore National Laboratory (LLNL) which dealt with electrons as the particles used to form the beams. SEESAW was terminated in 1972 due to the projected high costs associated with implementation as well as the formidable technical problems associated with propagating a beam through very long ranges in the atmosphere. One of the major difficulties in beam propagation is the bending of the ray. Sub-atomic particles, no matter how small, have substance or mass, and as such are affected by gravity and magnetism. As such, at long ranges, one cannot shoot a bolt in a direct line, but rather has to take into account the earth's magnetic field, which
bends the rays. As the earth's magnetic field is in a constant state of flux, aiming the weapon is extremely difficult.

Department of Defense interest was reinstated in 1974 when the United States Navy initiated the Chair Heritage program, again involving an electron beam development program. The Navy perceived an application to defense of ships against all forms of attack by aircraft and missiles. The Chair Heritage program differed in two ways from the DARPA program, which effectively increased its probability of success. First, the required range of the electron beam would be significantly shorter than for the SEESAW mission. Second, the beam was intended for point defense rather than for defense of large land areas, and the associated costs would therefore be lower. In 1977, the Chair Heritage program was shifted from a weapons to a technology base program due to a realization that major technological uncertainties still remain. In 1979, at the urging of Congress, the Chair Heritage program was transferred to DARPA.

In 1979 the United States Army initiated a second distinctly different path by starting the exoatmospheric, neutral beam program which is directed toward producing beams of electrically neutral hydrogen atoms. The Army also began a separate program to demonstrate "proof of principle" of a collective accelerator concept for producing high current ion beams. In fiscal year 1981 these programs were transferred to DARPA.

For several years the United States Air Force has been funding basic research on particle beam technology directed to several topics, including collective acceleration, propagation modelling and target effects. The Air Force has also been involved in developing new accelerator technology and propagation analysis for atmospheric applications, as well as initiating studies in areas that may be important for space applications: ion sources, beam control and power. The Air Force intelligence activities have also provided key information on Soviet technology efforts in particle beam development.

Beginning in fiscal year 1981, and continuing to the present day, DARPA has total responsibility for assuring technical compatibility of all United States military efforts. Due to the extremely high cost of any work in this area, and due to the poverty of funding for the programs ($49.6 million for 1984 — absolute peanuts in any military budget) two realities of this area of study have become clear: that DARPA is essentially the only organization involved in this work in North America; and secondly, that they are getting nowhere very, very, quickly due to lack of funding, with no weapon system even at the concept stage, much less on the horizon.

However, President Reagan has made public pronouncements on the use of "space age technology" as a nuclear deterrent, and so one is obligated to consider what he might mean, even if it is only hypothesis.

**Beaming Particles**

There are three key components of a hypothetical particle beam weapon system: the source of the beam — the beam generator — consisting of a particle accelerator and its associated supply of electrical power, energy storage and conditioning. The accelerators are similar to those used in research in elementary particle physics except that currents in the beam are much higher. The elementary particle research devices such as the two mile long Stanford Linear Accelerator have been widely publicized as "atom smashers." Second, there is a beam control subsystem to aim the beam at the target and determine that the beam has hit the target. Lastly, the particle beam weapon must have a fire control subsystem which acquires all the targets that need to be engaged, selects the one to engage, and tells the beam control subsystem where to look to find it. Then the fire control system decides when the target has been destroyed and designates the next target. These fire control functions do not differ materially from those of fire control subsystems for other more familiar weapons.

An appreciation of the damaging effect of highly energetic particles striking an object can be seen in many ways. The most easily visualized is the damage lightning can do when it strikes a tree or a house. In high energy physics, experimenters have long been aware of the ability of the highly energetic particles produced by atom smashers to penetrate into materials. As the beam penetrates, it transfers some of its kinetic energy to the material and, in addition, generates secondary radiation in the material which can also disable the target. If there are enough particles in the beam hitting the target, the rapid transfer of energy to the material cannot be dissipated by the material. Thus, the beam can cause a hole to be burned or melted into the material, or a fracture from thermal stresses as a result of the rapid deposition of energy. A third example of effects can be taken from discoveries in the early days of space flight. Energetic charged particles generated largely by the sun are trapped in the earth's magnetic field thereby forming the "Van Allen" belts. These natural particle beams require spacecraft designers to build shielded and resistant satellites if flights in or through these belts are to occur without damage to such "soft" components as computers or electronics.

As such, one can envision a weapon based on a stream of highly energetic particles that travel at or near the speed of light. This stream of particles would penetrate the metal skin of the target, transferring a large fraction of the energy in the beam to the target. Initially, as the beam enters the target it would damage electronic components and as the beam continues to deliver energy to the target, ignite fuels and explosives and/or create holes in the target ripping it apart.

In warfare, therefore, the theoretical beam weapon shares several "attractive" attributes with other forms of beam weapons (eg. high energy lasers. See "Military Lasers," ETI, November, 1983) in handling target tactics and scenarios that stress the capabilities of missiles and guns. For example:
Particle Weapons

— near speed of light delivery of destructive energy on line providing the earth’s magnetic field is compensated for.
— potentially large numbers of engagements before exhausting the available “magazine” as the “bullets” of a particle beam are, in effect, generated by the electrical power input to the beam generator.
— instant penetration through the skin of the target to destroy or disable key internal components or to ignite fuels and explosives.

If particle beam weapons are feasible, their use of energetic particles as the “bullets” offer two characteristics that reinforce those attributed to beam weapons in general, specifically:
— Since the particles can pass into the target and damage internal components without first burning a hole in the skin, the dwell time of the beam on the target could be quite short, even on targets that are hard to penetrate using other beam types. Keeping dwell times short allows the weapon to defeat the tactic of closely spacing the targets to saturate the weapon system.
— unlike laser weapons, the particle beam weapons can penetrate clouds and rain, giving the potential for an all weather weapon.

Raygunomics

It is recognized by all and sundry, with the possible exception of President Reagan, that particle beam technology is in the very early research and exploratory development phase, with fundamental issues of feasibility still to be resolved. How this is to be done, bearing in mind the high cost of experimentation and the low funding levels, is not explained. There is an enormous gulf between the technology required for fulfillment of the conceptual payoffs and the state-of-the-art. At this stage of development, DARPA has only

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defined generic "missions" to use as a basis for setting the technology goals. The key variables in these generic missions are range required, particle types used in the beam, and whether the deployment is in the atmosphere or outside of it. Conceptually, defensive applications (e.g., defense against attack by ballistic missiles, other types of guided missiles, and aircraft) are the logical underpinnings of their current research. This research is divided into a number of distinct sub-sections:

**Beam Generators:** The beam generator produces the intense, high energy beam of particles that represent the "bullets" of the particle beam weapon. Broadly speaking, charged particles are injected into an electric potential or voltage which gives the particles a "push" and accelerates them to high speeds. Kinetic energy is added to the particles by an amount equal to the product of the charge on the particle and the potential drop. Kinetic energy is proportional to the mass of the particle and increases with the velocity. Since the mass of the particles is small, their velocity is near the speed of light. The power of an accelerator beam is the product of the current in the beam and the potential drop. For example, an electron or proton has a charge of 1.6 x 10^-19 coulomb. A potential drop of a million volts will provide it with an added 1.6 x 10^19 joules or one million electron volts (MeV) of energy. One ampere of current (6.2 x 10^14 particles per second) falling through a potential drop of a billion volts has a power of a billion watts. Finally, if the billion watt system is pulsed for a millionth of a second, the pulse energy is 1000 joules or the approximate energy equivalent of raising a 750 pound weight one foot.

U.S. Department of Defense efforts can be broken down into two major thrusts: charged particle beam accelerators suitable for use only within the atmosphere, and accelerator systems capable of producing energetic neutral atomic beams for use only outside the atmosphere.

**DARPA** efforts are currently being placed on the development of high-current, moderate energy charged particle accelerators which are suitable for propagation experiments. A major part of the funds applied to the particle beam effort during the funding years 1983-84 is devoted to accelerator development. The main efforts in this line is the Experimental Test Accelerator (ETA) and the Advanced Test Accelerator (ATA) which are experimental electron accelerators.

The ETA program was completed in 1981. Since then it has been used as a testbed for ETA technology. It is currently being used to examine techniques for modifying the risetime and radial profile of electron beams to improve their propagation stability. The current program, ATA, produces 50 MeV electrons with a current of 10 kiloamps. It was completed in 1981 and is now used regularly, though only at low power. After the electron beam from ATA has been diagnosed, sometime late this year, it will be used for beam propagation experiments. These experiments, DARPA hopes, will provide the necessary information to show that electron beams can propagate stably for useful distances in the atmosphere.

A smaller effort is being pursued under the Neutral Particle Beam Program ("White Horse") at the Los Alamos National Laboratory to develop the accelerator technology required to accelerate negatively charged ions to high energies. Once high energy is achieved the excess electron is removed, thereby forming a neutral beam which can be directed over long distances in the vacuum of space. Experimental demonstration of the feasibility of generating low-divergence neutral particle beams forms the major element in this program.

The balance of this element in the DARPA program is directed toward investigation of various exploratory accelerator concepts that could have significant impact on the feasibility of this weapon system. The object of these investigations is to take advantage of new technologies which might become available for accelerator construction. An example is the Radial Line Accelerator program at the Air Force Weapons Laboratory, co-funded with the Department of Energy. The radial line accelerator concept proved feasible with the successful completion of RADLAC I in mid-1980. Presently, efforts are underway to examine a new class of transmission line accelerators in RADLAC II, scheduled for completion later this year. The RADLAC program is complementary to the DARPA ETA program in terms of the intended program physics experiments.

**Beam Control/Point Tracking:** Beam control subsystems for charged and neutral particle beams present technology requirements that are beyond the present state-of-the-art in all cases and without a technology...
Particle Weapons

basics in some cases. All are being addressed with a nominal development effort. The Air Force is primarily responsible for nuclear beam control and tracking. Efforts include design of a system that can sense the beam and provide the necessary extraction and pointing.

Prime Power and Conditioning: Efforts in prime power and conditioning are primarily associated with high repetition rate switching which is capable of handling the high currents used in accelerators suitable for charged particle beam weapon applications. Other efforts involve high voltage pulse forming networks, experimental capacitors and dielectrics, and new materials for high density storage systems.

Propagation of Charged Particle Beams: A major goal in the program is to provide experimental proof that charged particle beams can propagate through the atmosphere with sufficient power, arriving at the target at the ranges needed for weapon system applications. The major part of this effort is being performed by DARPA with experimental facilities at the ETA and/or ATA accelerators. The joint Air Force — Sandia National Laboratory RADLAC accelerator is used in this program.

Material Interaction, Damage and Effects: To understand what it takes to make a particle beam an effective weapon, research and experimentation is required to determine the interaction of particle beams with materials and components, the damage that results and the effects on target capabilities. The goals of the program for material interactions are to provide early assurance, as yet unforthcoming, that the beam can do lethal target damage. Should this element in the program succeed, the data generated will also define the beam power levels needed to assist in defining the R and D objectives in accelerator and pulsed power technology.

Star Wars

In summary, the objectives of the particle beam technology program are to determine feasibility and to develop the critical technologies required once feasibility has been demonstrated. The great majority of the available funds in the next few years will be devoted to building accelerators that can generate the high-current, high-power beams essential to verifying existing theory and that, according to DARPA, predict that beam propagation will be adequate for weapon feasibility.

As for the Soviets, their efforts are judged to be larger than that of the United States. It is particularly in the area of accelerators for fusion applications, and to have been in progress much longer. However, according to Department of Defense sources, there is no direct correlation between Soviet particle beam work and weapons related work.

Putting it all together, one therefore finds that there are half a dozen major problems to be solved, that some of the problems do not even have a theoretical basis for solution, and that funding for the research is small. Moreover, the Department of Defense, and perhaps President Reagan should have had a chat with his military advisors before embarking on a “Star Wars” scenario.
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IN THE EARLY DAYS of stereo, the channel separation used to be enormous. Stereo effect records can still be found in second hand shops that demonstrate this; a table-tennis match recorded so that the sound ping-pongs from side to side, or trains that pass through the middle of the house and so on. And anyone who has an early Beatles album in their collection will know that the vocals sound out from one side and the instruments from the other!

In those days, however, separate speakers were a novelty and the maximum separation of the speakers in a 'hi fi' radiogram was on the order of a few feet; then, they needed all the stereo separation they could cram onto a record. Since then the physical separation of the speakers has increased, while the electronic separation has decreased. So much so that many modern recordings sound like they were made by dedicated members of the "Back-to-Mono" club.

Since one cannot easily pull the walls apart to get increased separation, speaker placement is generally a matter of compromise between acceptable channel separation and the size and shape of the listening room. There is an alternative, however. It is possible to electronically separate the channels during playback, and this has the same effect as physically spacing out the speakers.

Sound in Space

A simple but effective circuit for achieving this is the subject of this article. But first to understand how it works it is necessary to understand the difference between stereo and mono sound recordings.

A mono recording is quite simple: all the information — vocals, instruments etc. — are all recorded as one signal. A stereo recording on the other hand consists of two signals or channels, which between them contain all the information. When the recording is made a single instrument can be placed on either the left channel or the right channel by means of a ‘Pan’ control (actually a simple potentiometer); rotating the Pan pot to the left puts all that signal onto the left channel, while rotating it hard to the right puts the sound on the right. Leaving the control set to the middle position effectively places the sound in the centre of the stereo image because the sound levels recorded on the left and right channels are equal.

However, the perception of stereophony also depends on the phase difference between the signals on each channel and it is this phase difference, which is imparted when the instrument is recorded in stereo, that enables us to further separate the sound after it has been recorded; if these phase differences are emphasised we can effectively increase the stereo separation.

Spaced Out Circuit

The diagram of Figure 1 shows the complete circuit of the Stereo Spreader. The Left and Right channel inputs are directly coupled by C1 and C2 to the non-inverting inputs of opamps IC1a and IC1b. To avoid the use of two batteries the op-amps are biased to half-supply by the resistive divider network R1, 2 and 3; capacitor C3 bypasses to ground any AC signal at the junction.

The op-amp outputs are connected together via the two halves of RV1 and R7. Now both are connected as non-inverting amplifiers, with feedback from each output to the respective inverting input via RV1a or b. Like all op-amps they will attempt to keep their inputs balanced by adjusting the output until the voltage fed back to the inverting input equals that present at the non-inverting input.

If the input to both op-amps is the same — i.e., in phase — then the outputs will be the same and normal op-amps action will apply. However, if the inputs are out of phase then the outputs will no longer be identical, so that part of the output of one op-amp will be coupled via RV1a, b and R7 to the inverting input of the other amplifier. This will then com-
pensate for the extra voltage and, in doing so, will produce a larger out-of-phase 'difference' signal.

The amount by which the difference signal is amplified is determined by RV1, which sets the amount of difference signal coupled from one op-amp to the other and therefore functions as a width control. A dual potentiometer is used here for convenience, so that it is not necessary to have to adjust two controls.

Construction
The Veroboard layout is shown in Figure 2 and, as long as the cuts are made in the right places and the board checked after assembly for unwanted solder bridges across tracks, there should be no difficulty in completing this part of the project.

For maximum flexibility the unit has been designed to be connected between the audio system’s pre-amplifier output and the power amp input. If these are not separate components in your system, check the back of the amplifier; most modern amps bring out the preamp outputs and the main amp inputs on pairs of phono sockets on the rear panel. If there is no way you can connect the Stereo Spreader between preamp and power amp, you will have to consider some other method of increasing your stereo separation, such as moving house or demolishing some walls! The only other alternative is to use the spreader with taped music only, in which case it can be connected between the recorder output and the amplifier inputs.

PARTS LIST

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<td>C3</td>
<td>100uF 10v radial electro</td>
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<tr>
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<td>TL082 dual BIFET op-amp</td>
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<tr>
<td>Miscellaneous</td>
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<tr>
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<td>sockets</td>
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</tr>
<tr>
<td>nuts and bolts</td>
<td></td>
</tr>
<tr>
<td>etc.</td>
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</table>

ETI
THE PRIMARY reason for AM stereo is to allow AM stations to compete more effectively with FM stations by improving the technical quality of the AM service.

Although AM does have a higher noise level than FM, the audio quality limitation is primarily due to poor radios. Most AM stations transmit audio to beyond 10 kHz, while FM signals go to 15 kHz, about the limit of human hearing. Further, virtually all AM radios use the antiquated envelope/diode detector process to recover modulation. Envelope detectors are inherently prone to the production of various types of distortion and other undesirable effects. One way that AM radios can be improved is by extending the bandwidth for better frequency response, and by replacing the envelope detector with a synchronous detector for reduced distortion.

However, the Federal Communications Commission (FCC) and the Canadian Radio and Television Commission (CRTC) regulations prohibit, for technical reasons, the expansion of bandwidths, negating the first of the above methods of improvement; while market constraints, e.g. the replacement of several tens of millions of radios currently owned by private individuals, preclude the changeover from envelope detectors. All that was left was the introduction of AM stereo.

The development and introduction of AM stereo over the past few years is positively Byzantine in its complexity, with many long knives being firmly embedded in many corporate backs. However, the end is in sight — we think.

Essentially, in the mid-70's, a series of researchers at Motorola in the US determined the basic groundwork for compatible quadrature modulation to provide AM stereo transmission and reception. Very quickly other companies — Belar, Harris, Kahn-Hasseltine and Magnavox — jumped into the fray and developed their own systems of reception. The difficulty was that the five systems were mutually incompatible, e.g. a radio would be able to receive the signal generated by only one of the five transmission methods. Needless to say, radio manufacturers were unwilling to produce radios under such a constraint, and an appeal was made to the FCC to choose a system which would be the industry standard.

In 1980, the FCC, for reasons best-known to themselves, chose the system developed by Kahn-Hasseltine. Much industry objection. The FCC reverses its decision. In 1982, the FCC authorizes stereo broadcasts on AM radio stations, but does not decide which competing technology will be the industry standard. Instead the Commission will accept whatever technologies it finds feasible and allows the marketplace (in line with Reaganomics) to decide which technology, if any, will become the industry standard. Belar essentially drops out, and Magnavox lowers AM stereo's priority in its R&D budget. Harris, Motorola and Kahn-Hasseltine commence battle, each claiming its system is better the others. Radio station execs are unsure of what to do, due to the high (40-120,000 dollar) cost of converting the station. Some jump the gun and buy Motorola, others Harris, some Kahn-Hasseltine. Most stay put and smile wanly, waiting for the dust to settle.

In late 1982, Motorola scores a coup — after extensive testing General Motors accepts their C-QUAM (Compatible Quadrature Amplitude Modulation) system for inclusion in all their car radios beginning in 1984. Much gnashing of teeth among radio station execs who have bought what now appears to be incompatible transmitting.
equipment, and champagne in Motorola's boardroom.

The Japanese Connection

Enter the Japanese — specifically Sony and Sansui. Knowing a good thing when they see one, they have independently and very quietly been developing a receiving chip which can accept all four seemingly incompatible transmission modes. Their introduction in January 1983 results in smiles of relief among radio station execs. Motorola execs revert to black coffee, as General Motors' lawyers sharpen their pencils and start hunting for contractual loopholes.

The Sansui/Sony chips are generically identical and apparently non-proprietary once one gets past the stage of whether the transmission mode is linear or non-linear. The Sansui chip, using a PLL synchronous detector instead of a conventional envelope detector, does this automatically, improving the signal to noise ratios and expanding frequency response. The Sony chip, while producing the same final result, requires the operation of a switch to determine whether the incoming signal is linear or non-linear.

It is not the purpose of this article to tiptoe through the bells and whistles of the four systems detailing their intricacies with the reverence of a Papal acolyte, but rather to demonstrate the essential differences between the two main thrusts in the technology — linearity and non-linearity.

A linear system is one in which superposition applies, that is, the system response to several inputs applied simultaneously is the same as the sum of the system responses when the inputs are applied individually. A non-linear system is one in which superposition does not apply.

Non-linear systems are best known from their application in the modulation of FM broadcasting. It has several advantages, the most prominent of which is its noise rejecting characteristics. However, such systems are dependent on spread spectrum systems (FM or PM, e.g. phase modulation).

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which trade improved noise rejection for increased RF bandwidth. But, in both the US and Canada, broadcast channel bandwidth allocation does not allow any increase in bandwidth — therefore no signal to noise ratio improvement can occur. In fact, non-linear AM stereo systems actually decrease the signal to noise ratio. Envelope modulation of L+R and angle modulation for L-R are present simultaneously and when the L-R signal causes the envelope modulation to reach -100% (known as pinchoff), the L-R signal disappears entirely, and no L-R modulation can be recovered.

Proponents of non-linear modulation techniques have not done so to improve signal to noise ratios (since non-linear systems actually degrade it) but rather as an attempt to obtain compatibility with envelope detectors already in place in most radios. The envelope detector, being a non-linear device, produces a distorted version of the amplitude modulation applied to it. By predistorting the I (In Phase) and the Q (Quadrature phase) elements of an AM stereo signal, it is possible to compensate for the distortion of the envelope detector. The predistorted non-linear AM stereo signal has an infinite number of sidebands, and in practice 3 or 4 orders of these sidebands are significant. To achieve compatibility with envelope detector receivers, all of the predistorted sidebands must be present at the receiver. However, this is not possible with most AM radios because the narrow IF bandpass filter the signal must pass through changes the relative amplitude and phase relationships. The only effect filtering at this stage will result in is distortion of components on mono as well as stereo receivers.

There are two types of compatibility distortion with non-linear systems: harmonic distortion and difference tone intermodulation distortion. Essentially, harmonic distortion occurs when a tone is transmitted and the high order sidebands of the linear system are altered in amplitude or phase. Difference tone intermodulation distortion on non-linear signals occurs in the opposite instance: in this case the desired signal is rejected by the radio, and only the accompanying low-frequency predistortion signal is received. This type of radio artifact is particularly objectionable since it is uncorrected within the radio output. This difference-tone distortion mechanism causes high-frequency programs to "throw trash" into the low-frequency range.

Proponents of this type of system argue that this type of distortion is of little significance since the program material generally contains high frequency energy. However, the use of pre-emphasis, that is, the boosting of higher modulated frequencies by as much as 20 to 30 dB (which is customary in radio transmissions for aesthetic reasons) severely aggravates the filtering related compatibility problem of non-linear systems’ signals.

As mentioned above, linear systems are those for which superposition applies. There are two important characteristics of linear systems, specifically that only one set of sidebands is produced on the same bandwidth as mono transmissions, and the linear product detection in both L+R channels means that there is no “noise burst” problem. The stereo signal consists of a monophonic sum signal with the linear addition of a double sideband suppressed carrier L-R signal in quadrature with the carrier. Since both signals have only one set of sidebands, and since addition is a linear process, the resulting stereo signal also has only one set of sidebands. No new frequencies are created by adding the two signal components. Further, this system is compatible with pre-emphasis and with synchronous detection (eg. non-envelope detection). A synchronous system has the advantage that there is no distortion due to receiver mistuning, selective fading or AM multipath (signals bouncing off buildings and suchlike).

Listening to the Radio

Yet, the proof of the pudding is in the eating, or rather, in this case, in the hearing by real live human beings. When defining the compatibility of the system with the human ear, the general evaluation is customarily considered to be low distortion. Distortion is usually taken to mean total harmonic distortion, or THD. The criteria for the evaluation of distortion should accurately represent the sensitivities of the human ear. For example, second order harmonic distortion is only objectionable when very large. On the other hand, difference tone intermodulation distortion is quite objectionable to the ear and should be of more consideration to the designers of these systems. In comparing non-linear with linear systems, it appears that while the linear systems have a higher THD than non-linear, that the bulk of the THD produced is in the second harmonic. In general, even order harmonic, such as second, fourth, sixth, etc., are less objectionable than odd-order harmonics. Further, the lower the order of the harmonic the less objectionable and audible it is. As such, second order harmonic distortion, which is produced more in the low order systems than in the non-linear systems is in fact less objectionable than the converse.

Essentially then, quadrature modulation is very appealing for application in AM stereo, since it will provide a chance for AM stations to compete with FM more effectively. Of the several systems in place, some simply reduce the amplitude of the L-R channel, thereby incurring a signal to noise ratio penalty. Other systems sacrifice the linearity of the system in various ways, forcing the transmitted envelope to be 1+L+R. These non-linear systems occupy excessive bandwidth, and in spite of the fact that the transmitted envelope is 1+L+R, radio IF bandpass filtering causes the non-linear system to generate distortion at the radio’s envelope detector.

While Australian and British stations have been experimenting with the various systems, it is in the United States where the majority of AM stereo stations exist, currently numbering about 100. Here in Canada, the CRCJ has licensed (for three months, extendable upon application) 15 stations across the country for AM stereo usage. The cost for the chips manufactured in the US, such as Motorola’s, is about $2 (US), with volume production coming on stream at the end of 1983. As for the Japanese, their chips and radios with the chips already built in are being snapped up as fast as they can produce them, with the bulk of shipments to this country being purchased by radio stations for testing and promotion purposes, though some are now beginning to appear in the shops. Further, when volume deliveries of the Japanese chips become available, it will be possible to retrofit them into radios already equipped for FM stereo reception with only minor auxiliary parts required, the major part being a ceramic resonator.
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DAC/ADC Filter Amplifier

ETI supplies the missing link for all your analogue to computer interface circuits. Design by C.D. Oddy.

ANYONE WHO HAS ever seriously experimented with analogue interfaces to home computers must be all too aware of the problems involved — if you’re not too careful you wind up with op-amps everywhere in a variety of hastily concocted amplifier, buffer, and filter circuits. And some of them won’t even be hastily concocted; some of them will have taken up a considerable amount of your precious time in the designing. What you need is a handy little unit that will perform any or all of the functions of amplifier, buffer, and filter at the mere twiddle of a control or two. What you need is the ETI filter-amplifier.

The filter-amplifier consists of two active blocks together with input and output buffers and switchable AC or DC coupling. The two active blocks are an amplifier with a variable gain of 0 to 100 and a plus or minus five volt variable offset followed by a low pass filter whose cutoff frequency may be varied over the range 16 Hz to 30 kHz and which may be switched out of circuit when not required. It can be used as an amplifier to match low level signals to the input of an ADC, as a buffer to correct mismatching of signal sources, as a filter to smooth the stepped output of a DAC, and for a multitude of other similar purposes.

Construction
Assembly is simplified by the use of a single quad op-amp for the four stages, and the only point to check on the PCB itself is that this IC and the two electrolytic capacitors (C2, C3) are inserted the right way around. Take care with the wiring of the potentiometers and switches, especially SW2 which has two separate elements whose wiring

Fig. 1 Circuit diagram of the filter-amplifier.
must be in agreement if the filter is to work correctly. Note also the wiring around RV4 and the connection between this potentiometer and SW2, as shown in Fig. 2.

We have not shown either a case or a power supply for this project because it is intended more as a building block than as a complete, self-contained unit. For the really serious experimenter there is obviously a lot to be gained from having several of these filter-amplifiers built into a single unit and equipped with a common power supply, whereas for others, one would probably be sufficient and could be powered from whatever you’re using to supply the ICs you’re experimenting with.

![Connection diagram](image)

**How It Works**

The input signal is passed via either C1 or SW1 to IC1a, the first stage of the TL074 quad BIFET op-amp, which acts as a buffer. The signal is then amplified by IC1b, the gain being set by R2 and RV3. RV1 and RV2 set the offset voltage, RV1 being preset such that the offset is zero when RV2 is in its centre click-stop position.

IC1c and its associated circuitry form a low pass filter of the second order Bessel type which is well suited to the smoothing of stepped signals, such as those obtained from DAC outputs. The filter’s cut-off frequency may be adjusted using RV4 and the range switch SW2, the latter selecting either 16Hz to 3 kHz (position 1) or 160 Hz to 30 kHz (position 2). The gain of the stage is set by R7 and R8.

SW3 selects either the amplifier or amplifier plus filter outputs via R9 and R10, the slight difference in their values being necessary in order to allow for the gain (approx. 2.3 db) of the filter. The signal then passes through IC1d, the final buffer stage, to the output.

---

**Parts List**

<table>
<thead>
<tr>
<th>Component</th>
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<tr>
<td>Resistors (all 1/4W, 5%)</td>
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<tr>
<td>R1</td>
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<tr>
<td>SW3</td>
<td>SPDT</td>
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</table>

PCB; case, input and output sockets, knobs, etc. as desired.
COMMODORE BUSINESS Machines has decided to enter the portable computer field by packaging the workings of the Commodore 64 in a sturdy metal case with one disk drive and a five-inch colour monitor. Add to this Commodore's software for business applications, and you have the makings of a product that could be aimed at the commercial or educational market. One of the strong selling points is compatibility with existing software and peripherals for the VIC-20 and 64. It has all the familiar functions: 64K RAM, sound synthesizer, sprites, 320 by 200 pixel graphics, and others. The suggested list price is $1200.

Since the Commodore 64 was reviewed in ETI, July, 1983, this review will look at applications of the SX64 rather than tinkering under the hood.

Starting with the packaging: you'll find the case good and solid, fairly heavy at 12.5 kg, and sized at approximately 43 by 43 by 13 cm. You wouldn't want to go jogging with it, but it sure beats doing a balancing act with a monitor, a power supply, a 64, and a drive.

A Quick Tour
Unlatch and remove the front cover; on the inside of it is a full-size keyboard with 66 keys identical to the regular 64. A cable plugs easily into the keyboard and awkwardly into the computer's underside, and the usual three-pin power cord plugs into the back. During your travels, you can fold the cords into a storage space above the disk drive; Commodore supplies you with a cord case evidently made from floppy disk boxes, but everybody will probably cram them into the computer anyway.

On the top of the computer, you'll find the slot and connector for ROM cartridges, and on the rear panel are the two joystick ports, the DIN plug for audio/video, the DIN plug for another disk drive, the expansion bus, a fuse, and the power switch. Current consumption is one ampere at 117 volts. In other words, it has exactly the same features as the regular 64, except for slightly higher power consumption because of the add-ons. There's a carrying handle which doubles as a tilt stand if you fight with two little latches that let it rotate around the front of the computer.

The Monitor
A five-inch monochrome monitor is hard enough on those of us who don't have microscopes for eyes, but a colour unit? Since it's safe to say that potential customers will at least view the screen before buying it, it isn't necessary to go on at length. Suffice to say that the apparent resolution is coarse, very coarse; passable for video games, but colour bleed makes text difficult to read.

A good giggle was to load in an 80-column word processor; the screen was just one big smear. It was usable on 40-column.

Some improvement can be made by changing the screen to black and white; this is POKE 53281,0 CTRL 2.
doesn’t work if your software uses its own colour control codes to reset the colour anyway. You can also fiddle with the display controls; behind the CBM logo on the right-hand side is a tiny door which conceals colour, tint, contrast, and so on. There’s also a volume control for the internal speaker.

The Disk Drive
The ordinary 64 can use a CBM 1541 disk drive with 170K of storage capacity and a serial in/out system. The apparent reason for this is to simplify the wiring for hooking several peripherals together (“daisy-chaining”); it keeps your carpet looking like a trawler net. The disadvantage to the minimum-wire serial system is that the data feeds in or out one bit at a time instead of one byte at a time — obviously everything runs much, much slower than, say, a CBM 4040 parallel drive.

Commodore has made the peculiar decision to incorporate this 1541-type drive into the Executive. This means that loading a program is painfully slow: a large program took two minutes to amble into the memory; compare this to the ten seconds or less taken by a Apple running CP/M. This waiting around isn’t exactly outrageous, and it’s acceptable for noodling about with games on a lazy Saturday, but it’s annoyingly tedious for any serious work with spreadsheets, word processors, or other business software.

The name Executive implies a business-oriented computer; the decision to use the serial drive proves a serious drawback in light of this. On the other hand, if you have to use up a warehouse full of 1541s...

CP/M?
CP/M is the famed disk- and file-handling system which does such a great job of almost instantly locating and loading any file from any disk with simple commands instead of all those frustrating syntaxes used by most small micros, including CBM’s. Commodore is not selling the necessary option in Canada, supposedly not until the spring, though they may have it together by the time you read this. Of course, CP/M would be slowed down considerably by the serial drive anyway.

Should you have been dead keen on CP/M, though, a Toronto firm called Computer Workshops Ltd. has come up with the necessary Z80/80-column additions. You can contact them at 465 King St. E., Unit 9, Toronto M5A 1L6. Keep in mind that 80 columns are not usable on the Executive without an external monitor.

CBM
Commodore Business Machines supplied the computer for review with no manuals and no further information. Several calls to the company produced two manuals; they were the same ones you can get for the ordinary 64, and they hadn’t sent any information about the disk operating system. If this wasn’t just an oversight, a lot of customers will be left puzzled, because there’s very little disk information in the computer manuals. We’re left wondering about things that are not apparent from staring at the computer itself. There’s a little Reset button near the video controls, for instance, and it didn’t seem to do much. Is it for those unfortunate moments when a program hangs up? You might well ask. We also received American press releases showing the Executive with dual disk drives, but the local public relations people said that only a single drive is now available. An interesting marketing approach.

A Trial Run
They did manage to include some optional software, though. The first was the Easy Script word processor. This consisted of a manual, a quick reference card, and two apparently identical disks containing the program. I’m not sure if the two disks were a mistake, but it’s nice to have a backup.

The main menu, which appears after two minutes of loading, asks you to configure the program to suit: text width, disk/tape, and printer type. The text width doesn’t change the size of the characters, but makes the screen into a window which scrolls over the text width you’ve selected (up to 240 columns).

The Easy Script is generally nice to use; the function keys become control codes, which is convenient. It has a few oddities, of course: calling up the disk directory munches through your text if
you didn’t happen to notice in the manual that you should have saved it to disk first, and the cursor commands are a bit cumbersome; end--of--text, for instance, is F1/G/E(R). This is needlessly complicated; software writers seem to think we’re a nation obsessed with touch-typing. There is also no automatic justification of the right margin; this has to be done manually with RETURNs; this is a real pain if you're used to auto wordwrap, but probably no bother if you’re coming into it from a typewriter or similar word processor.

The Manager
If you have a computer with everything included in the case, it would make sense to have a business program with all sorts of applications included on the disk. Such a program is The Manager, supplied for the 64 by Commodore. It consists of a manual and a disk.

After a two minute load, you’re presented with a menu which lets you format a disk, among other things. Good thing, too. There must be a lot of 64 owners out there phoning their dealers to say that they can’t find anything in the manual.
This time I try Enter/Edit (another reload) and ask for the “Checkbook” sample. Up it comes, after a while. The graphics are quite good, with a ch-e-q-u-e displayed in various colours. Amounts (and other information) could be typed into the appropriate spaces. The same goes for the Christmas List program, which lets you store away all sorts of facts about your friends and enemies.

OOPS! That RESTORE key got nudged again...abandon ship.

Calc Result
This spreadsheet arrived with the Executive, but is actually published by Handic Software. It’s a manual, a disk, and a ROM cartridge. After loading for some minutes, it asks you which of a number of languages you’d like; I mean humanoid ones, not computer languages. This is a nice touch. Now it would like you to make a backup disk and a work disk; this takes over ten minutes, although you only have to do it once.

Thereupon, a sheet is spread for you. F5 gives you a help menu, though I could only get it with the original disk; it didn’t seem to be present on the backup. The screen layout itself is nine columns by 21 rows; the cursor can be scrolled over the entire spreadsheet, which is 63 by 254. Co-ordinates can be typed in to move the cursor directly to any spot, and there’s auto-repeat on the cursor as well.

You can enter either labels or numerals, and even enter an arithmetic operation; Calc solves the arithmetic and enters the value at the desired co-ordinates. Then you can specify various operations from one spot to another: B2-B4, 40% of B2, and so forth. The final result (the famous “bottom line”) appears; if you change any value, the whole works is re-calculated.

Like most spreadsheets, Calc Result lets you split the screen. To show the entire sheet would require a screen the size of the wall; splitting it lets you look at rows and columns that are too far apart for the normal display.

This program was friendlier than the others, and didn’t require interminable loading whenever you wanted an option. The Restore key didn’t dump things for you, either.

Summary
Keeping in mind that the colour screen is a wee bit hard on the eyes, and that the disk drive will appeal only to those who’ve never used anything but cassette loading, $1200 suggested list isn’t too steep a price to pay to get all your peripherals in one box, especially if dealers discount the price.

The software sent to us seemed too slow and cumbersome for any business use except the occasional weekend dip into the bytes. However, if you’re intrigued with the Commodore 64 and its many possibilities, and want the convenience of portability, this is a good unit to investigate.

Quick Reference
Commodore Executive SX-64
Mfg: Commodore Business Machines
Price: $1200
CPU: 6510
RAM: 64K
User RAM: 38K
Software Included: BASIC 2.0
Screen Format: 40 x 25
Graphics: 320 x 200
Colour: 16 available
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The March issue of CN! for example, will be featuring the usual double helping of articles and features that CN!'s readers have come to know, including “Receiving Radio Teletype on an Apple”, an interface to allow one to pick up news wire transmissions from a short wave radio and decode them into text. There will also be “Machine Language on VIC-20’s and 64’s” to help Commodore system users to explore the inner workings of their machines. “A Look At Dedicated Terminals” and “An Introduction to Multi-user/Multi-tasking Systems” will dispel a lot of the myth and confusion that’s built up around this aspect of high end business micros.

March will also bring our valuable reference feature “The Directory of Computer Stores”. It will be a complete listing of the computer retailers across Canada ... invaluable for you for months after the magazine has been released.

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ETI—MARCH—1984—35
SOONER OR later you may have the bad luck to become the owner of a misbehaving computer. The trouble will usually be one of two types: the first is the permanent fault, which is always there, and the second is the maddening glitch that only appears after you’ve typed in three pages of programming; it may only change one word to something else, or it may ignore the keyboard and stubbornly refuse to let you do anything but pull the plug.

The first type of trouble is the easiest to find, because the fault will hold still to be measured. The second is a little more challenging. In either case, where do you start?

How To Maintain And Service Your Small Computer is a very good place indeed. This 208-page softcover is a wealth of information for someone reasonably familiar with basic test equipment, such as multimeters, logic probes and scopes. At least you now have a place to start, once you get under the hood and realize that there’s a lot going on in there, and all of it looks hard to find.

The book begins with some sensible advice concerning the quality of the power line, disk care, and interference with your TV set. There isn't much point in crunching through your computer's innards if the fault lies with 5 KV spikes coming down the hydro wires.

The next part gets a bit silly; there's a guide to what basic meters look like, the types of screwdrivers, and even advice on how to lift heavy objects safely. Oh, well; you never know. Perhaps someone who doesn't know how to lift heavy objects can find their way through computer circuitry. Once you get through this, though, there's an excellent guide to the structure of the circuitry; the function and in-circuit testing of passive components is well covered and explained. Occasionally it's just a bit misleading for the inexperienced: advice is given on how to test the power switch with an ohmmeter, but the authors recommend turning off the power instead of pulling the plug. It might be smoky ohmmeter time.

Chapter Seven has an excellent explanation of the computer's layout, and how you go about using a scope and logic probe to trace the signals. Naturally, a book this size can't possibly include all the different makes and their methods; the reader is expected to have some ingenuity at translating the generic approach into his computer's specific hardware.

Chapters Eight to Twelve cover repair and maintenance of disk drives, printers, cassette systems, and power supplies. There's also a quick look at modems, monitors, and game controllers. As with the computer explanations, the reader shouldn't expect an exhaustive study, but the topics are explained in enough depth to provide a good beginning.

The book concludes with five appendices of which the IC pinouts and troubleshooting charts will be the most useful. In summary, you may not become skilled in computer repair overnight, but you'll at least know what to look for and where to look for it.

Bill Markwick
1984 Directory of Computer and Electronics Stores in Canada

Memo: Re Directory Listings

The information in this directory is that supplied by the companies themselves in most cases and in every case where a product range is mentioned.

Moorshead Publications accepts no liability for errors, however caused. Where an error is significant, we will publish corrections in a future issue.

Since the number of outlets for ETI and CN! is changing daily, references to this should be regarded as a guide only, though it should be correct at the time of going to press.

If your company is not listed, or if the listing is inaccurate, please send details to: Directory Listing, Moorshead Publications, 25 Overlea Blvd., Toronto, Ontario, M4H 1B1.
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<td>IBM, Compaq, DEC Digital, Hyperion, Apple, Neuma,</td>
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<td></td>
<td>Computer Circuit Ltd.</td>
<td>ETION 733 Richmond St., London, Ont. N6Z 3H2 (519) 672-9370</td>
<td>MO</td>
<td>Wang, Hyperion, Apple, Commodore, North Star hardware. All types of peripherals, software, media, and supplies. Full service support centre including training.</td>
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<td>Computer Innovations</td>
<td>CN 149 Dundas Street, 5th Floor, London, Ontario N6A 1G3</td>
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<td>Forest City Surplus Ltd.</td>
<td>ETI 781 Dundas St., London, Ont. N5W 2Z6 (519) 436-0233</td>
<td>Inventory changes constantly and includes all kinds of new manufacturers surplus computer hardware.</td>
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<td>Radio Shack Computer Centre</td>
<td>1298 Trafalgar St. London, Ont. N5Z 1H9 (416) 453-0511</td>
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<td>The Information Connection</td>
<td>199 Queens Ave., London, Ont., N6A 1J1</td>
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<td>Markham</td>
<td>Carolian Systems Inc.</td>
<td>290 Don Park Rd., Unit 11, Markham, Ont. L3R 2V1</td>
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<td>Compucentre</td>
<td>Markville Shopping Centre, 5000 Hwy. 7, Markham, Ont., L3R 4M9</td>
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<td>Computer Connection</td>
<td>7311 Woodbine Ave., Markham, Ont. L3R 3V7 (416) 492-9500</td>
<td>Specialists in small and medium business applications. Products Include the full Apple product line, Zenith, Epson, Onyx, plus a full complement of software and accessories.</td>
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<td>Electronic Surveillance Corp.</td>
<td>270 Ferrier St., Markham, Ont., L3R 2Z5 (416) 475-0550</td>
<td>MO CAT $10.00</td>
<td>Computerized security systems, computerized burglar &amp; fire alarm systems, computer-aided monitoring systems, computer-aided dispatch systems, computerized control systems, S-100 &amp; ESC interface systems, computer-aided design systems.</td>
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<td>Quester International Inc.</td>
<td>7270 Woodbine Ave., Markham, Ont. L3R 1A4 (416) 475-8044 or 496-1211</td>
<td>Commodore and Olivetti computers, software and peripherals. Design and develop application software for small to medium-sized businesses.</td>
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<td>Teo Computers &amp; Peripherals</td>
<td>275 Steelcase Rd., Markham, Ont. L3R 1G3 (416) 474-9372</td>
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<td>Versacom Systems Inc.</td>
<td>431 Alden Road, Unit 1, Markham, Ont. L3R 3L4</td>
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<td></td>
<td>Teo Computers &amp; Peripherals</td>
<td>275 Steelcase Rd., Markham, Ont. L3R 1G3</td>
<td>FREE</td>
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<td>Wheat Systems Corp.</td>
<td>13 Old Colony Rd., Manotick, Ontario KOA 2N0 (613) 692-2425</td>
<td>FREE</td>
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### Craig Office Products Corp.
- 329 Waverley St., Ottawa, Ont. K2P 0V9 (613)238-5500
- MO
- IBM PC, Victor, Seiko, IFS, Oki Data, Xerox.

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### Micro Comp Ent.
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- MO
- Commodore - IBM - Victor - total supplies - diskette, ribbons - paper - buy & sell used equipment.

### Pickering
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- MO
- Commodore - IBM - Victor - total supplies - diskette, ribbons - paper - buy & sell used equipment.

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400 North Christina, Sarnia, Ont., N7T 5W4
Radio Shack Computer Dept.
293 Bay St., Station Mall, Sault Ste. Marie, Ont., P6A 1X3
Sault Office Machine

ScARBorough
MD Data Systems
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Comp-CAN
1477 Hunt St., Sarnia, Ont. N7S 3MB
Keelan's Audio & Video
460 North Christina, Sarnia, Ont., N7T 5W4
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Radio Shack Computer Dept.
293 Bay St., Station Mall, Sault Ste. Marie, Ont., P6A 1X3
Sault Office Machine
<table>
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<tr>
<th>Directory of Computer Stores</th>
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<table>
<thead>
<tr>
<th>Store Name</th>
<th>Address/Contact Details</th>
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<tr>
<td><strong>Sudbury</strong></td>
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<tr>
<td>A &amp; B Office Equipment</td>
<td>2148 LaSalle Blvd, Sudbury, Ont., P3A 2A7</td>
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<tr>
<td>ComputerLand</td>
<td>Unit #19 Cedar Point Plaza, 1984 Regent St. S., Sudbury, Ont. P3E 5S1 (705) 522-3663</td>
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<tr>
<td>Computerline</td>
<td>ETI 74 Cedar St., Sudbury, Ont. P3E 1A5 (705) 674-3315 MO</td>
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<tr>
<td><strong>Xerox</strong></td>
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<td>63 Church St., St. Catharines, Ont., L2R 3C4</td>
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| **Toro**                        |                         |
| Active Components               | ETI 14 Carlton St., Toronto, Ont. M5B 1K5 (416) 977-7692 MO CAT FREE |
| Complete electronic components and computer products needs. Texas Instruments, Motorola, Fairchild, General Instruments, Anadex printers, Hewlett Packard, IBM computers, Star microcomputers, Micro Pro, Tandy, Tandon drives, Apple IIe, Macintosh, Hyperion, Compaq, Neima Personal, DEC Rainbow 100 and 100 plus. |

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| **Allan Hamm Sales Ltd.**       | 2389 Bloor St. W., Suite 300, Toronto, Ont. (416)769-1773 |
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| **Compucentre**                 | 2nd Floor, Ste. 207, Eaton Ctr. Galleria, Toronto, Ont. M5B 2H6 (416)961-5978 |
| **Compumart**                   | TEC 505, 218 Yonge St., Eaton Centre, Toronto, Ont., M5B 2H6 (416)596-1300 |
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| **Drumondville Computer Systems** | 14 Ontario St., St. Catharines, Ont. L2R 7M3 (416)894-7776 Zenith business systems and software, full service centre (Zenith). |
| **Radio Shack Computer Centre** | 290 Glendale Ave., St. Catharines, Ont. L2T 2L3 (416) 227-2020 |
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Computer Book & Supply
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Computer Innovations
Corporate Offices: 401 Bay St., Suite 2115, Toronto, Ont. M5H 2Y4 (416) 864-0808

Computer Generation
1722 Avenue RJ, Toronto, Ontario M5M 3Y6

Computer Innovations
4800 Yonge St., Toronto, Ont., M2N 5M7

Computer Innovations
44 Bloor Street East, Toronto, Ont. M4W 3H7

Computer Innovations
176 Yonge Street, Toronto, Ont. M5C 3L7

Computer Innovations
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Comspec Systems
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C.P.A. Ltd.
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Compul-Group
2255 Sheppard Ave. E., Unit 136, Toronto, Ont. M2J 4Y1 (416) 494-2900

Computer Junction
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Computer Mail Inc.
37 Mutual St., Toronto, Ont. M5B 2A7 (416) 362-2753

Computer Shack
567 King West, Toronto, Ont. M5V 1M1 (416) 595-5777

Computer Workshop
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Delphi System Group Inc.
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Delphi Systems Group
71 McCaul St., Toronto, Ont., M5T 2X1

Dimensional Business Systems
199 Bay St., 4th Floor, Toronto, Ont., M5J 1L4

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First Byte
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Gladstone Electronics
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MO CAT FREE


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Iris Computer
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Magsparta Information Systems Inc.
519 King St. W., Toronto, Ont. M5V 1K4 (416) 596-8911

MO CAT FREE


Marketron Corp.
1240 Bay St., Suite 802, Toronto, Ont., M5R 2A7

M & W Computer Store
407 Queen St. W., Toronto, Ont. M5V 2A5

Micronet Computer Centres Ltd.
465 King St. E., Suite 208, Toronto, Ont. M5V 2B4

Microcomputer Store
2632 Yonge St., Toronto, Ont. M4P 2J5 (416) 487-5551

MO

Products carried: IBM, Apple, Digital, Altos computers, Epson printers, C-1TOH printers.

Micronet Computer Centre
1 First Canadian Place, Station A, Concourse Level, Box 193, Toronto, Ont., M5X 1A6

Micron Dist.
409 Queen West, Toronto, Ont. M5V 2A5

Auby Mandell Software & Books.

Microcomputech Electronics Ltd.
ETI/CN
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MO

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Micro Marketing Can.
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35 Queen St. W., Toronto, Ont. M5V 2A4

Perfect Electronics
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Radio Shack Computer Supply
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Radio Shack
1938 Avenue Road, Toronto, Ont. M5M 1A1 (416)987-3997

MO CAT FREE

Radio Shack
111 Ave Rd., Toronto, Ont., M6C 1W2

Radio Shack
1 Dundas St. W., Eaton Centre, Toronto, Ont., M8B 1G8

The T. Eaton Company Ltd.
290 Yonge St., Toronto, Ontario

Xerox
110 Eglinton Ave. E., Toronto, Ontario, M5H 2R3

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TAF Computer Systems Ltd.
275 Sable St., Toronto, Ont. M6M 3K8 (416)245-4916

MO CAT FREE

North Star microcomputers (ADVANTAGE and HORIZON), associated peripherals, software packages and supplies.

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T. Eaton Company Ltd.
1 Dundas St. W., Eaton Centre, Toronto, Ont., M8B 1G8

The T. Eaton Company
290 Yonge St., Toronto, Ontario

Xerox
335 Bay St., Toronto, Ont., M5H 2R3

CompTree Store
of Hamilton inc.
13 Bold Street
Hamilton, Ontario
L8P 1T3 Canada
527-8733

Uxbridge
Uxbridge Electronics
Box 364, 75 Brock St. N., Uxbridge, Ont., L0C 1K0

Vanler
Radio Shack
150 Montreal Rd., Les Galeries Vanier, Vanier, Ont., K1L 8N2 (613)748-1462

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Waterloo
Abacus Computers
180 King St. S., Waterloo, Ontario, N2J 1P8

Compulier Software Inc.
380 King St. S., Waterloo, Ont. N2J 2Z3 (519)681-6011

Compulier Software Inc.
500-22 King St. S., Waterloo, Ont. N2V 1N8 (519)746-1880

MO

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Compupro Micro Systems
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Affordable Computers
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Research and development in software for the utilization in network and multiluser environ- ment with UNIX operating system and on database environment.
<table>
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<th>Directory of Computer Stores</th>
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<td><strong>Dollard-Des-Ormeaux</strong></td>
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The Birds of Babel: Satellites for the Human World

SOMEBWHERE FAR off in the lonely, airless reaches of space is a lot of lonely hardware. A hundred years ago, nobody would have dreamed of how obsessed modern man (or at least some modern men) would become with putting hardware up where he couldn't reach it. Back then they would never have understood that nowadays it isn't cool anymore to take a piece of hardware in hand and use it. You have to use it by remote control, ideally from as far away as possible. Ask anyone who's ever used a TV remote control. Who knows, maybe it was the invention of the car bomb that started this mad stampede away from modern technology. In any event, the epitome of this phenomenon is the geostationary communications satellite, orbiting 22,300 miles above the heads of those who live on the equator. Up here in Canada, we use the things, but we don't have them hanging over our heads.

The Birds of Babel: Satellites for the Human World is a book about satellites — how they've developed, what they've done, what they're likely to do in the future, and why. The book focuses primarily, but not exclusively, on communications satellites, those that are in that 22,300 mile orbit. It tells how the orbit was discovered by science fiction writer Arthur C. Clarke, who was the first to realize that at a certain critical height, a satellite would orbit exactly as fast as the earth turns, and would thus appear to stay over a fixed point on the planet, capable of broadcasting signals to about a third of the earth's surface. A reasonably non-technical description of how satellites work is given, along with descriptions of some of the companies, such as INTELSAT, that are putting satellites up there.

The book is quite well written, and very readable. It is by no means a technical manual, and does not try to give you enough information to build your own satellite, but rather looks at them from the point of view of those who use them and benefit from them — or in other words almost everyone in a country like Canada, or in fact in virtually every country around the world. In other words, it's a good book.

Anthony DeBoer

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ETI—MARCH—1984—37
Frequency Counter

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A FREQUENCY COUNTER is a very useful piece of test equipment as it can be used for troubleshooting both analog and digital circuitry. The design shown in this article uses readily available TTL and CMOS ICs for measuring frequencies from below 10 Hz up to 10 MHz.

This counter differs from others in several important areas:

The gating times are generated by a crystal controlled clock which is very accurate and will not drift significantly with time as opposed to the more commonly used monostable multivibrator.

When measuring low frequencies (below 10 KHz), the input signal is multiplied by up to 100, thereby reducing the reading time to one second or less.

This frequency counter automatically updates its display about three times every second (approximately once per second on the 100 Hz range), and a latch is used to provide a flicker-free display of the input frequency.

A block diagram of the frequency counter is shown in figure 1. The input signal, after being conditioned by the input circuit, is "anded" with the gating pulse and the resulting signal is fed to the counter, which counts the number of high to low transitions at its input. This count is then latched and displayed on the four

Schematic diagram for the frequency counter.
Power supply schematic for the frequency counter.

seven-segment displays. If, for example, the gating pulse is set at one second and the input frequency is 2259 Hz, the counter will 'see' 2259 high to low transitions in the one second gating period and thus 2259 will be displayed. Refer to the "How it Works" section for a more detailed description of the frequency counter.

Construction

The printed circuit boards shown in this article are recommended to speed construction and reduce the possibility of wiring errors; however, wire-wrapping techniques could be used if desired. The following construction details apply when the suggested Hammond case is used with the printed circuit boards.

Start by assembling the large board. First solder in the jumpers; there are 24 of them. Use a short piece of insulation to insulate J1 from the crystal leads. Next, solder in the IC sockets and the power supply parts. Be sure to attach heatsinks to both the bridge rectifier and the regulator before soldering them in place. Neither silicone nor mica insulators are required for either heatsink. The remaining parts can now be soldered in place. Pay attention to the orientation of the tantalum capacitors, transistors and diode.

Before inserting the ICs, temporarily apply power to the board and check the power supply. The input to the regulator should be 9V or greater (depending on the transformer used) and the output of the regulator should be 5V, within half a volt. Check each IC socket separately for +5V and ground (excluding the two 16 pin header sockets and the DIP resistor socket). Pin 7, 8, or 9 should be ground and pin 14, 16, or 18 should be +5V for 14, 16, and 18 pin sockets respectively. If there is a problem, check that all of the jumpers are installed correctly. With ground and +5V at each IC socket, power should be removed and the ICs inserted; note the proper orientation of each one. The DIP resistor obviously can go in either way; if, however, it is not available eight 33 ohm resistors bent for 0.4 inch insertion can be used instead.

With the main board completed, the display board can be assembled next. Solder in the ten jumpers and the two IC sockets, then insert the four displays using the parts overlay and figure 2. The decimal point of each display should be on the bottom half of the display board. Next a six inch sixteen pin header cable should be made as indicated in figure 3 and connected to both boards as shown on the parts overlay.

Having completed both printed circuit boards, the rotary switch can be wired. Initially it should be tested with an ohmmeter to verify that each position of each pole is operational; this may save considerable time if troubleshooting is required later. Use table 1 to connect an eleven inch piece of ribbon cable to the switch.

At this point in construction the unit can be fully tested. With the rotary switch and the display board connected properly, apply power to the main board. If all is well the display will read '000' or '001' and the leftmost digit will be extinguished. If there is a problem, refer to the troubleshooting section, otherwise continue testing. Apply about a 2V peak to peak signal to the input for each of the six ranges; in each case the display should read out the correct frequency. For frequencies above approximately 200 KHz, a TTL equivalent signal should be applied to J17, bypassing the input circuit since it only responds to signals below this frequency. With a signal applied to J17, the frequency counter should read signals up to 9,999 MHz. Also check that the leftmost digit is extinguished when it is zero, and that the decimal point moves according to the rotary switch position. If all ranges work properly, continue construction by mounting all parts inside the cabinet.

The front and back panels which come with the cabinet are over three inches high and should both be cut down to the two inch mark and appropriate holes drilled. The display board and bezel can be mounted to the front panel followed by...
<table>
<thead>
<tr>
<th>Problem</th>
<th>Area of Problem</th>
<th>Possible Cause of the Problem</th>
</tr>
</thead>
<tbody>
<tr>
<td>all four digits extinguished</td>
<td>display driver</td>
<td>- DISP1-DISP4 properly oriented</td>
</tr>
<tr>
<td></td>
<td>circuitry</td>
<td>- ribbon cable connected properly</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- DIP resistor in place</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- IC11 properly inserted</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- Q6,Q7,Q8 properly inserted</td>
</tr>
<tr>
<td>leftmost digit extinguished</td>
<td>first-digit blanking circuitry</td>
<td>- Q3,Q4,Q5 correctly in place</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- R8-R11 proper values</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- proper insertion of IC9,IC10</td>
</tr>
<tr>
<td>decimal point not on</td>
<td>decimal point</td>
<td>- proper connection of ribbon cable</td>
</tr>
<tr>
<td></td>
<td>switching circuitry</td>
<td>- wiring of switch 1, pole 'e'</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- R19 (DIP resistor) okay</td>
</tr>
<tr>
<td>display on, but unit won't count on any range</td>
<td>input circuit or clock circuit or reset circuit</td>
<td>- input BNC connector</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- wiring of switch 1 poles 'a', 'b' and 'd'</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- R3,C3,R4,C4 proper values</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- IC6 inserted correctly</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- operation of the clock and clock divider circuitry (IC1-IC5)</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- proper insertion of IC9-IC11</td>
</tr>
<tr>
<td>operational on all but the lower three ranges</td>
<td>input frequency multiplier</td>
<td>- switch 1, pole 'c'</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- all parts associated with IC7 and IC8</td>
</tr>
<tr>
<td>operational on all but the 10 MHz range</td>
<td>input frequency divider</td>
<td>- switch 1, pole 'a'</td>
</tr>
<tr>
<td></td>
<td></td>
<td>- proper wiring of IC3b</td>
</tr>
</tbody>
</table>

Table 2. Troubleshooting guide.

**MEMORY SPECIAL**

<table>
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<tr>
<th>Part Number</th>
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<tr>
<td>4116 150nS NEC</td>
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<tr>
<td>4864 1x64K Hitachi 150NS</td>
<td>$8.99</td>
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<tr>
<td>2708 1Kx8</td>
<td>$5.55</td>
<td></td>
</tr>
<tr>
<td>2716 2Kx8 Hitachi, NEC</td>
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<tr>
<td>2732 4Kx8 Hitachi, NEC</td>
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</tr>
<tr>
<td>2764 8Kx8 Mitsubishi</td>
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<tr>
<td>6532</td>
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<tr>
<td>2532</td>
<td>$7.65</td>
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<tr>
<td>2114 200NS (Hitachi)</td>
<td>$2.25</td>
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</tr>
<tr>
<td>2126-2</td>
<td>$7.59</td>
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</table>

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<table>
<thead>
<tr>
<th>Part Number</th>
<th>Description</th>
<th>Quantity/Price</th>
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<tr>
<td>Z80</td>
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<tr>
<td>Z80A NEC</td>
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<td>$6.75</td>
</tr>
<tr>
<td>6502A 2MHz</td>
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<td>$6.99</td>
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<td></td>
<td>$6.50</td>
</tr>
<tr>
<td>6845 CRT Controller (Hitachi)</td>
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<td></td>
</tr>
<tr>
<td>6522</td>
<td></td>
<td>$6.99</td>
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<tr>
<td>6088</td>
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<tr>
<td>8746D NEC</td>
<td>Peripherals</td>
<td>$24.50</td>
</tr>
<tr>
<td>8741 NEC</td>
<td></td>
<td>$32.50</td>
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<td>8251 NEC</td>
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<td>8257-5 NEC</td>
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<td>LS323</td>
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<tr>
<td>LS508</td>
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<td>LS11</td>
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<td>LS132</td>
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<td>LS368</td>
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</tr>
<tr>
<td>LS86</td>
<td></td>
<td>.55</td>
</tr>
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</table>

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the rotary switch, BNC connector and the power switch. The back panel only has the power connection. The main board is mounted inside the cabinet with the input molex connector adjacent to the BNC connector. Mount the transformer and wire the power switch followed by the BNC and power connectors. Finally, plug the two 16 pin headers into the appropriate sockets.

The frequency counter is now complete and should be tested again. If input signals greater than 200 kHz are frequently going to be measured, the input circuit should be redesigned to respond to signals up to 10 MHz. Alternatively, a switch could be used to take the input circuit out of the signal path in which case the input signal would have to be TTL compatible.

**Use**

The frequency counter is very easy to use. An appropriate signal is applied to the input BNC connector, and the corresponding frequency is read off the display. If the recommended input circuit is used, the input signal should be less than 200 kHz with an amplitude greater than two volts peak to peak. When taking a reading, the highest range should be selected first and lower ranges selected until the left display is lit; this avoids misreading the counter when there is an overflow. The lower three ranges of the counter may take a few seconds to stabilize after a signal is applied while the phase-locked loop locks onto the input signal. On all but the 100 Hz range, the display is updated about once every 300 ms; the 100 Hz range, having a gating time of one second, is updated once every 1.3 s or so.

**Troubleshooting**

If the frequency counter does not function properly, first check the power supply for +5 V. Both circuit boards should be inspected for solder bridges and all IC leads should be checked for proper insertion. If the problem(s) still persist, use table 2 as a guide in troubleshooting the unit.

---

**PARTS LIST**

Resistors (all 1/4W 5% unless otherwise specified)

<table>
<thead>
<tr>
<th>Resistor</th>
<th>Value</th>
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<tbody>
<tr>
<td>R1,2</td>
<td>270R</td>
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<tr>
<td>R3,8,9,20</td>
<td>4k7</td>
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<tr>
<td>R4</td>
<td>390k</td>
</tr>
<tr>
<td>R5</td>
<td>18k</td>
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<tr>
<td>R6</td>
<td>47k</td>
</tr>
<tr>
<td>R7</td>
<td>10k</td>
</tr>
<tr>
<td>R10</td>
<td>1k</td>
</tr>
<tr>
<td>R11</td>
<td>10R</td>
</tr>
<tr>
<td>R12-19</td>
<td>33R DIP resistor (Bourns 4116R-001-330 or equiv.) *</td>
</tr>
<tr>
<td>R21</td>
<td>4M7</td>
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Capacitors

<table>
<thead>
<tr>
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<tr>
<td>C1</td>
<td>2200u 25 V electrolytic</td>
</tr>
<tr>
<td>C2</td>
<td>10n ceramic</td>
</tr>
<tr>
<td>C3,8,9,10,11</td>
<td>100n ceramic</td>
</tr>
<tr>
<td>C4</td>
<td>1u tantalum</td>
</tr>
<tr>
<td>C5</td>
<td>4u7 tantalum</td>
</tr>
<tr>
<td>C6</td>
<td>50p silver mica</td>
</tr>
<tr>
<td>C7</td>
<td>2u2 tantalum</td>
</tr>
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</table>

Semiconductors

<table>
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<tr>
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</thead>
<tbody>
<tr>
<td>Q1-Q8</td>
<td>2N3904 or equiv.</td>
</tr>
<tr>
<td>D1</td>
<td>1N4148 or equiv.</td>
</tr>
<tr>
<td>BR1</td>
<td>1.5A, 200 V bridge rectifier with heatsink (W02M or equiv.)</td>
</tr>
<tr>
<td>REG1</td>
<td>5V, 1A regulator with heatsink (LM7805)</td>
</tr>
<tr>
<td>IC1,10</td>
<td>74L594</td>
</tr>
<tr>
<td>IC2,3,4,5,8</td>
<td>74L5390</td>
</tr>
<tr>
<td>IC6</td>
<td>74L5221</td>
</tr>
<tr>
<td>IC7</td>
<td>MC14046B</td>
</tr>
<tr>
<td>IC9</td>
<td>74L8132</td>
</tr>
<tr>
<td>IC11</td>
<td>74C926 (National)</td>
</tr>
</tbody>
</table>

* These parts are available at Newark Electronics, 271 Attwell Drive, Rexdale, Ontario, M9W 5B9.

The 74C926 is a National part available at Zentronics and other National distributors.
The printed circuit layouts for the main board and display.

Parts overlay for the frequency counter.

Figure 4. Header cable diagram.
The frequency counter can best be understood if it is broken down into five separate sections: the power supply, the input circuit, the frequency multiplier, the clock generator and divider, and the counter/display driver.

Power Supply
The secondary voltage from the transformer is full wave rectified by BR1 and filtered by C1; it is then fed to the regulator where it is regulated down to 5 VDC. The output of the regulator is used to drive all of the ICs and the unregulated 12 VDC (or so) is used in the input circuitry. C9 improves the stability of the regulator.

Input Circuit
Input signals are fed to the main board via the BNC connector on the front panel. C7 removes any DC component from the input signal, and D1 clips off the negative half of the signal. Q1 and Q2 amplify the input to a square wave which is fed to the input of IC9a; the output is connected to both a decade divider (IC5b) and switch 1 pole 'a'. While on the 10 MHz range the divided signal from pin 3 of IC5 is used; on all other ranges the direct output from IC9a is used. The output from S1a feeds both switch 1 pole 'b' and the frequency multiplier.

Frequency Multiplier
An MC14046B phase-locked loop (PLL) is configured to multiply the input frequency by 10 on the 10 KHz range and by 100 on the 100 Hz and the 1 KHz range. The signal from S1a is buffered and inverted by IC10a and fed to the input of the PLL (pin 14). This input signal is compared to the output of the decade dividers IC9a and b, and a comparison signal is generated at pin 13; this signal is filtered by C5, R6, and R7 and drives the voltage controlled oscillator input (pin 9). The VCO output at pin 4 is buffered and inverted by IC1d and drives both the decade dividers and the input of IC9b. The VCO adjusts its frequency until the input signal and the comparison signal from the decade dividers are equal. At this point the PLL is locked, and the output frequency is either 10 or 100 times the frequency at pin 14, depending on the position of the rotary switch.

Clock Generator and Divider
IC1a, b, c, and the crystal are configured to oscillate at a frequency of 5 MHz. The 74LS390 decade dividers reduce this frequency to produce the required gating times of 0.01s, 0.1s, and 1.0s. Switch 1 pole 'd' selects which of the gating times will be used; the following explanation assumes a gating time of 0.1s. At the end of the 0.1s, the output of IC4b will swing from low to high, triggering one half of IC6 (a dual one-shot timer). This one-shot generates a 'high' pulse approximately 300us wide (determined by the time constant of R3 and C3) at pin 13. This pulse transfers the output of the counter into a latch and triggers the second one-shot. A 300ms pulse is produced at pin 5, resetting both the decade dividers and the counter, but not the contents of the latch. With the decade dividers reset, another gating pulse is generated and the contents of the latch are updated. The gating pulses from S1d are inverted and 'anded' with the conditioned input signal to produce the 'count' signal. A timing diagram showing the clock signals is given in figure 4.

Counter/Display Driver
The 74C926 is a complete counter, latch, and display driver. The counter, latch, and reset signals are connected to IC11 and the appropriate display signals are generated. R12-R18 limit the current to the displays and Q6, Q7, and Q8 drive the common cathode of DISP4, DISP3, and DISP2 respectively. DISP1 is blanked when it is zero, and therefore requires some extra circuitry. Note that DISP1 is zero when segment 'f' is on and segment 'g' is off. Q3, Q4, IC10d and e, and IC9c and d decode this condition and blank the display. Switch 1 pole 'e' selects the proper decimal point to light.
Our Notebook topic is that much-misunderstood beast, the switched mode power supply. P.S. Wilson of International Rectifier gives a step-by-step explanation of the various types and design examples.

The term 'switching mode power supply' is used to describe DC-to-DC converters and AC-to-DC converters which operate on a switching principle. Using switching techniques, voltage step-up and voltage inversion can be achieved, as well as the more common voltage step-down function. The advantages of using switching techniques over a linear solution are the reduction in the size of components (such as power transformers and output filter capacitors) by operating at high frequency, and dramatic improvements in efficiency, since the power elements are either fully turned 'on' or 'off' and do not operate in the linear mode. The disadvantages of switching mode solutions are increased noise and radio frequency interference (RFI) which is generated during the switching transitions. Circuit complexity is increased, as in addition to the control circuit, a power switch, rectifier, high frequency transformer or inductor and drive circuitry is required.

Switching mode solutions are, however, cost-competitive with linear power supplies in off-line applications at and above the 100 W level. Switch mode power supplies are also used at lower power levels in DC-to-DC converters where there is a special requirement such as high efficiency, for example in solar energy conversion, or small size for mobile communications equipment, or computers.

Basic Principles

The circuit and waveforms in Figure 1 illustrate the basic principle of the switching mode supply by comparison with a linear regulator. The circuit configuration shown is for a voltage step-down conversion. When switch SW1 is closed, the input supply voltage is applied to the inductor L1, and current flow in the inductor will rise with a ramp waveform, charging capacitor C1 and also supplying the load connected at the output of the supply. When SW1 is opened (equivalent to turning off a semiconductor device), the inductor current diverts into the rectifier, DL. The voltage at circuit node 'P' falls instantaneously to a rectifier forward voltage drop.
below the 0 V line, and the current flow in the inductor follows a negative ramp waveform. The power supply load is now supplied both from the inductor and from the output capacitor, \( C_1 \). When SW1 again closes, D1 becomes reverse biased and the inductor is again connected to the input supply. In the steady state condition, the positive volt-second product applied to the inductor must balance the negative volt-second product applied when the rectifier conducts. The voltage at the output of the supply is regulated by controlling the on/off ratio, or duty cycle of the switch SW1. Because the switching element is either 'on' or 'off', the power loss is small and the efficiency of the supply approaches 100\%. Comparison with the linear regulator (Fig. 1b) shows an efficiency of approximately \( \frac{V_{OUT}}{V_{IN}} \).

Figure 2 illustrates how, by rearranging the circuit elements SW1, \( L_1 \) and D1, voltage step-up and voltage inversion can be achieved. Provided that the current flowing in \( L_2 \) does not fall to zero between the conduction phases of SW1, the circuit configuration in Figure 2a can be said to provide a 'non-pulsating' output current. This feature allows low output ripple voltage to be achieved. The configuration shown in Figure 2b, however, will exhibit a 'pulsating' output current as the inductor current is diverted from the output when SW1 closes. The input current flow, however, can be arranged to be non-pulsating, so reducing the ripple voltage on the input supply. The voltage inverting circuit, (Fig. 2c) has pulsating current waveforms at both input and output terminals.

To overcome this apparent restriction on operating mode, transformer-coupled circuits can be used. The voltage conversion achieved is then defined by the transformer turns ratio and the polarity of the output rectifiers. Figure 3 illustrates the most common circuit configurations in use today. In addition to increasing flexibility, the transformer-coupled solution offers the option of an isolated output supply.

Figure 3a shows a transformer-coupled circuit configuration analogous to the voltage step-up circuit in Figure 2b. The dots against the transformer winding indicate their polarity. SW1 and D1 conduct simultaneously. During the switch 'off' time, current flow in \( L_1 \) is diverted through a second rectifier, D2. The purpose of the third winding on the transformer is to reset the magnetic core of the transformer during the switch 'off' time. If this was not done, the magnetic core would become DC-biased and may saturate, resulting in poor performance (low efficiency and high pulse currents in primary winding and SW1).

Figure 3c, the push-pull converter, is again analogous to the circuit in Figure 2a. The difference between this circuit configuration and the forward converter shown in Figure 3b, is that the transformer is biased bidirectionally by switches SW1 and SW2 which conduct alternately. Consequently, the 'reset' winding shown in Figure 3b is not required. The output filter components \( L_1, C_1 \) operate at twice the switching frequency, allowing some size reduction. Each switching device (SW1, SW2) passes only one-half of the output current divided by the transformer turns ratio, \( n \). Consequently, this solution may be preferred to the solution shown in Fig. 3b at higher power levels (greater than 100W).

Figure 3d illustrates a type of push-pull converter commonly used in off-line applications. Its main advantage, apart from the automatic resetting of the transformer core, is that the maximum
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Fig. 3. Transformer-coupled switching mode circuits. (a) Flyback converter. (b) Single ended forward converter. (c) Push-pull converter. (d) Half bridge circuit. (e) Full bridge circuit.

Voltage seen by either switch does not substantially exceed the input supply line voltage. Consequently, 200 V switches can be used when working directly from the rectified 120 V mains supply. Capacitor C2 prevents DC biasing of the transformer core which may otherwise arise through asymmetry in the switching waveforms of SW1 and SW2. Capacitors C3, C4 effectively divide the supply to the transformer by two.

Finally, Figure 3e represents a further modification to the basic forward converter in Figure 2a. The capacitors C3, C4 in the previous figure are replaced by two more switches; SW3, SW4. DC magnetization of the transformer core is prevented by capacitor C2. The full supply voltage is now applied across the transformer primary as switches SW1 and SW4 and then SW2 and SW3 close simultaneously. The maximum voltage applied to any of the switches will not exceed the supply voltage significantly. This 'full bridge' configuration is used in high power switching power supplies where the size and cost of capacitors C3, C4 to replace the switches would be prohibitive. The same circuit configuration is used to drive reversible DC motors. To be continued
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The previous article explained how the computer handles input and output of data, but how many of us read and write in binary? This month Owen Bishop explains the operation of more user-friendly interfaces.

Quite a number of electronic circuits produce their output in analogue form, and it is beyond the ability of a computer to read such data unless it is first converted to digital form. Examples of devices with analogue outputs are electronic thermometers, pressure transducers, audio amplifiers (for speech recognition etc.) and indeed any device which produces an output voltage varying over a predefined range. Even a simple carbon pot can be included in this list. The games controller of the Apple II, for example, uses a 150k potentiometer (there is also a push-button connected to a memory-mapped latch, but this is a digital input). The position, or setting, of the potentiometer is the analogue quantity to be measured. The computer has a quadruple timer IC, and when the games controller is plugged in to the computer board, the pot becomes part of the RC circuit of the timer. When the MPU is to read the setting of the controller, it first triggers the timer (the trigger input is memory-mapped) then measures the length of pulse produced. It does this by reading the memory-mapped output over and over again, counting how many times it reads ‘high’ until it eventually reads a ‘low’. The number of ‘high’ reads is approximately proportional to the angular setting of the control knob. The analogue-to-digital conversion is crude and far from linear, but certainly good enough for its intended application.

A-to-D Conversion...

Many computers have on-board A-to-D ICs such as the National Semiconductor ADC0801 (Fig. 2). This converts any input voltage in the range 0 V to 5 V, to digital output in the range 0 to 255 (00 to FF in hex). The heart of the IC is a chain of resistors in series, with 5 V across the ends of the chain.

Internal logic controls CMOS analogue switches which switch resistors into or out of the chain, so producing a voltage (Vc) which can range from 0 V to 5 V in 256 steps. At each stage, a comparator matches the output of the chain against the analogue voltage (Vin). The largest resistor is switched in and out first, to determine if the input is less than or greater than 2V5. Then the next largest resistor is switched in and out to narrow the possible range to within 1V25. At each stage the chain output and analogue input are matched more and more closely. After eight attempts, the closest match will have been found. The logic signals which have produced the match, are then used to set the eight output buffers to one of the 256 possible combinations, which can be read as a byte (0 to 255) by the computer. This IC converts the voltage with true linearity, with an accuracy of half a step in 256 steps and takes only a few hundred microseconds to do so. If we want greater accuracy, there are similar A-to-D ICs with a 12-bit output. Note that it is the converter which does the work; the MPU only has to read the result. We do not need to write software to instruct the MPU to measure pulse lengths as with the

![Fig. 1. Analogue-to-digital conversion for a games controller.](image1)

![Fig. 2. Simplified block diagram of an A-to-D converter such as the ADC0801.](image2)

---

**Fig. 1. Analogue-to-digital conversion for a games controller.**

**Fig. 2. Simplified block diagram of an A-to-D converter such as the ADC0801.** $V_{\text{REF}}$ is an on-chip or external reference voltage. $V_{\text{c}}$ is the output voltage from the resistor chain, which is compared with $V_{\text{in}}$, the analogue voltage which is to be converted.
games controller. This saves time and simplifies programming.

Most A-to-D ICs have ways of altering the span of the input range, so that voltages from, say, 0 V to 2 V produce the full-scale output range, 0 to 255. In addition, you can adjust the offset so that, for example, you obtain a reading of 0 when the input is 10 V and a reading of 255 when the output is 12 V.

As might be expected, an A-to-D IC is a sophisticated circuit and is correspondingly expensive. If an application requires several analogue inputs, and high conversion time is not of paramount importance, the inexpensive 507C IC provides linear conversion in 1 ms with seven-bit resolution. This has a resistor ladder (Fig. 3), and the counter supplies current to each resistor in binary sequence. The op-amp adds the currents and the result is that the output ramps down a voltage from 0.75 x supply voltage to 0.25 x supply. If the enable input is high, the comparator gives a high output whenever the ramp voltage exceeds the analogue input voltage; thus the length of time the output is low, is a measure of the analogue voltage (Fig. 4). The MPU can find this time by using a program like that described for the games controller.

---

**Fig. 3.** Simplified block diagram of a simple A-to-D converter, the 507C. VREF is an on-chip or external reference voltage. VIN is analogue voltage to be converted; VOUT is the square wave output.

**Fig. 4.** The input and output voltages of the 507C A-to-D converter. The computer measures $t$, which is proportional to VIN.
.. And The Converse Conversion

If a peripheral needs an analogue signal to control it (eg. controlling the speed of a motor), we need circuits which can convert the digital output of the computer into its analogue equivalent.

The ZN425 D-to-A converter makes use of an R-2R ladder (Fig. 5). The switches are under the logical control of the eight-bit input. As the count increases, the output voltage increases in proportion (see panel). To drive a low impedance device, the output must be buffered by an operational amplifier.

This IC may also be used to A-to-D conversion, using its binary counter. The counter is clocked by an external pulse generator, and as it counts the pulses, switches controlling the R-2R ladder are closed and opened in a binary sequence. The output is a staircase ramp of 256 steps. An external amplifier compares the output from the ladder with the analogue voltage which is to be converted. When the ramp output equals the analogue voltage, the output from the amplifier inhibits the clock. At this point, the logic output, which controls the switches, can be read as an eight-bit equivalent of the analogue input.

Screens And Printers

The way most owners receive information from their micros is through the screen. This may be a domestic TV set or a monitor unit specially designed for the purpose. Those whose main interest lies in arcade games, adventures, and the like, usually need no more than the screen, but anyone with an interest in programming soon finds that the screen alone is not enough. There is the tedium of copying long listings of favourite programs from the screen, and the frustration of being able to see only a few lines of program at any one time. Sooner or later, the serious programmer adds a printer to the system. This month we shall deal with both these ways of receiving information from the micro.

The fact that the technology of high-speed (for the time) printing was already available in the form of teletype machines, lead to the early mainframe computers having a printer but usually no screen. We are reminded of this early use of teletypes by some of the curious control codes which abound in the ASCII character set (see last month’s article). The screen of the early mainframe computers, if any, was often a CRT in which the X and Y deflection plates were under the direct control of the computer. It was a kind of high-grade oscilloscope, with the computer using the electron beam to 'draw' an image on the screen. This was suitable for displaying charts and graphs, but not much use for text.

![Fig. 5. Block diagram of the ZN425E D-to-A converter. V_REF is an on-chip or external reference voltage. The amplifier is not on the chip, but is an external op-amp (eg. 741).](image)

![Fig. 6. How text is displayed on a TV screen.](image)

![Fig. 7. Enlarged view of the first few letters shown in Fig. 6.](image)
With a modern computer, the electron beam scans a rectangular area on the end of the CRT, in the same way as a TV set. The field (or raster) consists of a large number of horizontal lines (the number varies according to the system) placed close together (Fig. 6). Each line is scanned in turn, from the top to the bottom of the screen. Most TV systems have an interlaced raster in which the beam first scans alternate lines down the screen and then returns to scan the ones between the first set. As the beam scans, its intensity (and hence the brightness of the glow produced on the screen) is modulated. If the beam is strongly modulated, so that it produces either bright light or none, we can use it to produce a textual display of good contrast. Figure 7 shows how a line of text can be built up from dots of light in seven successive scans. These are followed by a number of lines in which the beam is blanked (off) to provide spacing between rows of text. This process is repeated all the way down the screen.

As Figure 7 shows, it is possible to build up well-defined alphabetic characters from a 7x5 matrix of dots. Numeric characters, punctuation marks and various other symbols can also be built up in this way. We will now look in more detail at what the computer has to do in order to produce such a display. Visual displays are a field in which microcomputer designers have felt themselves free to use their inventiveness. Consequently, there are almost as many ways of producing the display as there are makes of microcomputer. Our discussion will therefore deal only with the main principles which are common to most micros.

### The Writing's On The Screen

Figure 8 shows the signal which is fed to the grid of the CRT to modulate the beam to produce scan 5 of Figure 7. The length of a single dot-producing pulse is of the order of 15 μS. This waveform is produced by a circuit like that shown as a block diagram in Figure 9. Most micros have what is termed a memory-mapped display. A certain block of memory addresses is set aside for holding the text. Normally it requires one byte of data for each character. For example, if the screen has 16 lines with 64 characters per line, the memory area must consist of 1025 bytes, or 1 kilobyte. Whenever text is to be displayed, the CPU stores the ASCII codes corresponding to each character in the appropriate memory cell. The video RAM, as this section of memory is usually called, is read in sequence by the video circuitry. Although video RAM can be addressed by the MPU.
Assume all switches are set to 0 V. The three left-hand resistors are equal to 2R and 2R is parallel, and so can be replaced by a single R resistor. Thus the four resistors shown are equivalent to 2R switched to 0 V. We can carry this reasoning all along the ladder, until we reach a switch that is not set to 0 V.

If all except SW7 are set to 0 V, and SW7 is set to V (the reference voltage), we can consider all resistors to the left as equivalent to a single 2R resistor, switched to 0 V. We have a potential-divider, and \( V_o = V/2 \). This corresponds to the expected output, since 128/256 = 1/2.

If all except SW6 are set to 0 V and SW6 is set to V, we can consider all resistors to the left to be replaced by a single 2R resistor, switched to 0 V.

\[
I_o = \frac{V}{3R}; I_1 = (V - V_i)2R; I_2 = \frac{V_i}{2R}
\]

By Kirchoff’s Law (sum of currents entering a point must equal sum of currents leaving a point):

\[
I_o = I_1 - I_2,
\]

\[
V_i/3R = (V - V_i)/2R - V_i/2R,
\]

giving

\[
2V_i = 3V - 3V_i - 3V_i,
\]

\[
V_i = 3V/8
\]

Now we can calculate \( V_o \):

\[
V_o = 2V_i/3 = V/4
\]

This corresponds to what is expected since 64/256 = 1/4.

Here both SW6 and SW7 are switched to V.

\[
I_o = (V - V_i)/3R; \quad \text{but } I_1 \text{ and } I_2 \text{ are as above.}
\]

\[
I_2 = I_o + I_1, \quad \text{so}
\]

\[
V_i/2R = (V - V_i)/2R + (V - V_i)/3R,
\]

which simplifies to \( V_i = 5V/8 \).

Now \( V_o = V_i + (V - V_i)/3 \)

\[
= 5V/8 + V/8 = 3V/4.
\]

This is what we expect, since 192/256 = 3/4.

like any other part of RAM, in some micros it has its own control signals and its own data bus to connect it with the video circuit, for most of the time it operates independently of the microprocessor. The ASCII code for each character in turn is transferred to the data latch. It is held there while the next code is being fetched. The output from the latch goes to a character generator IC. This is a special kind of ROM (see Designing Micro Systems, ETI December 1983), which converts an ASCII code into the corresponding pattern of dots. It has inputs from the latch to tell it which character is to be generated, and from the synchronizing circuit, to tell it which one of the seven lines of dots is to be generated. It has five outputs that indicate which dots are to be displayed as white and which will be black.

The output from the character generator is converted from parallel to serial form by the shift register. This produces a train of pulses, some high, some low, like those shown in Figure 8. These are fed to a video mixer circuit where line synchronizing pulses and frame synchronizing pulses are added. The combined video signal is then passed through a buffer circuit to the monitor.

Such a signal is not suitable for sending to a domestic TV set. The TV is expecting a VHF carrier signal from an aerial, modulated by the video signal. So, if a TV is being used, the signal from the video mixer must first go to a modulator. This produces a VHF carrier signal (usually on channel 3 or 4) with the video signal imposed on to it. When the TV receives this signal, it demodulates it, recovering the original video signal that it then uses to produce the display. The additional processes of modulation and demodulation inevitably lead to distortion; consequently, the resolution obtainable on a TV is inferior to that obtained on a proper monitor screen. A TV is acceptable when there are 40 or fewer characters per line, but when there are 60 characters or more, the use of a monitor is much to be preferred.

Next month, we look at the format of serial and parallel data transmission for communicating with peripherals.
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For galactic Hitch-hikers everywhere, the ideal travelling companion for your Electronic Thumb and Guide to the Universe. Designed and developed by Rory Holmes.

NOWADAYS THERE are many scientists of the paranormal who will assure us that a person is able to influence and control matter by the direct action of his mind alone. They call this effect psychokinesis; it's akin to the claims of the familiar ardent gambler who is convinced he can influence the throw of a dice, or the spin of a roulette wheel, to come up with his lucky number.

ETI, with an open mind, decided to design a machine which would give a clear indication of any such psychokinetic action directed on it. We thought that a 16 ton weight linked to a movement detector might perhaps give the skeptics an unfair advantage and so decided that the slightly smaller and rather more capricious nature of the electron might be more amenable to the delicate influences from the psyche. In fact, the electron itself does not have to be influenced, but rather those forces involved in producing the random movement of thermally and electrically excited electrons. These forces are completely unknown and unpredictable; that's why they're called random! It's this nature of randomness and its ultimate source which has fascinated people for years; hence the favourite pastimes of staring into the fire or watching the ocean waves.

In our machine, the random motions mentioned above are linked electronically to control the movement of a dot of light round a circle of LEDs. This dot of light obviously shouldn't rotate too far in any one direction, but should dither around — both clockwise and counter-clockwise — keeping the same position on average. The light moves at regular clock intervals and at every movement it has, theoretically, a 50-50 chance of going in either direction.

We called it the Infinite Improbability Detector for two reasons. First, in the absence of any assumed psychokinetic intervention from a lively mind, it is incredibly unlikely that it will rotate consistently in one direction. Second, we're Douglas Adams fans. The probability against continuous rotation is, in fact, two to the power of 16 to the power of the number of revolutions. Therefore, if this occurs, don't panic — we have proof of an external and paranormal force.

It is hoped that if psychokinetic ability does exist, then it should certainly be able to bias our machine to rotate in one direction or another.

In use the operation is simple — if a little indefinable. There is a rate control for setting the overall speed of movement of the light and an on-off switch. After switching on and setting the desired speed, the rest is really up to you. The idea is to concentrate on the moving dot of light in order to make it consistently rotate in one direction.

ETI would be delighted to hear from any readers who can demonstrate impressive results.

Construction

It isn't difficult to obtain a neat and smart appearance with this project (see the photos of our prototype). There are only two front panel controls, making interwiring and assembly completely straightforward. The only point to watch during construction is the mounting of the LEDs; these should be left until the rest of the PCB is assembled.

Solder in all the other components, following the overlay guide. It's easier to start with the links and sockets first, then the passive components. Leadout wires to the pot (RV1) and battery supply can also be connected at this stage.

After assembly, the LEDs (LED1 to 16) can be mounted with their anodes towards the centre of the circle, but do not solder them at this stage — the LEDs must be fitted into their front panel holes to provide the right uniform height above the PCB before they are soldered. Take a piece of tracing paper and mark the exact hole positions from the copper track pattern. This can then be transferred to the inside of the case front panel and the positions centre-punched before drilling the holes to a width exactly the same as the LEDs. The sloping front Vero box we used has a detachable aluminum front panel which makes this an easy process. The LEDs are then fitted into the holes and the correct board clearance height can be found before they are soldered up.
We found that after soldering, the good fit of the LEDs was quite enough to hold the PCB in place. Two 9V batteries were fixed inside the case with adhesive pads, and battery clips were used to connect them up to the switch and the PCB. The potentiometer and switch are mounted through two holes on the front panel, after wiring on the front panel, and after wiring them up from the overlay diagram the device can be tested.

**NOTE:**
IC1 IS 4069B
IC2 IS 4029B
IC3 IS 4051B
IC4 IS 4013B
Q1 AND Q2 ARE 2N3904
LED1-16 ARE 2mm RED OR YELLOW LEDs

Fig. 1. Component overlay for the PCB. The LEDs are arranged in a circle with the anodes to the inside.

Fig. 2. Circuit diagram of the Infinite Improbability Detector.
Infinite improbability Detector

Switch on and set the speed control to its lowest rate; if all is well the illuminated LED should move backwards and forwards round the circle in a random manner with no preferred direction. If it does rotate in only one direction, then there is either a construction blunder, or the portion of the Universe occupied by the detector has passed through a region of Infinite Improbability.

PARTS LIST

Resistor (all 1/4 W, 5%)
R1,3 1MΩ
R2,4,8 10kΩ
R5,7 4MΩ
R6 1kΩ
R9 330kΩ
R10 47R
R11 150R
R12 47kΩ
R13 3kΩ

Potentiometers
RV1 1MΩ logarithmic pot

Capacitors
C1 330nF polycarbonate
C2,4 100nF ceramic
C3 10nF ceramic
C5 470uF 16V axial electrolytic

Semiconductors
IC1 4069B
IC2 4029B
IC3 4051B
IC4 4013B
Q1,2 2N3904
LED1-16 2 mm red or yellow LEDs

Miscellaneous
SW1 miniature toggle switch
PCB; two 9V batteries plus clips; case.

HOW IT WORKS

IC2 is a four bit up/down counter, set to count in binary mode and thus having 16 count states provided on its four output lines. These binary output lines are decoded by IC3 and Q2 to give a one-of-16 output for driving LED1 through LED16.

IC3 is a CMOS one-of-eight analogue decoder, whose outputs are connected to the anodes of two sets of eight LEDs. The select lines on pins 9, 10 and 11 will take one pair of anodes to positive via R10 for each address. When the D output of IC2 is logic low the cathodes of LEDs 1 to 8 are held low, thus illuminating one of them and Q2 will be switched off. When D goes logic high the reverse takes place, with Q2 now turned on to provide a ground for the cathodes of LEDs 9 to 16. Thus each count state will correspond to one LED being illuminated.

Assume for the moment that the up/down control on pin 10 of IC2 is kept high (the 'up' mode); then, for each clock pulse applied to pin 15, IC2 will move to its next count state from 0 through to 15. Each LED in the circle will in turn illuminate to indicate that particular count, so producing the effect of a rotating dot of light.

However, the up/down control is made to continuously change in a random fashion and at a much faster rate than the clock period. The timing diagram of Fig. 2 shows the clock as a square wave. The CMOS counter IC2 only changes state as the clock moves from low to high and the direction of its counting can only be changed when the clock is high. Thus the state of the up/down control when the clock goes low will determine the count direction when it next goes high again. The dot of light will move randomly either clockwise or counter-clockwise on each clock period. If, say, the light moves three places clockwise in succession, this is equivalent to getting three heads in a row from a coin toss.

The random logic level generator is based on Q1, a transistor selected to provide a good noise output when connected in its base-emitter breakdown mode. This produces genuine random noise due to the completely unknown and unpredictable movements of excited electrons in the semiconductor material. The noise voltage so generated is amplified by IC1c and d. These are unbuffered CMOS inverter gates which are biased into their linear amplifying mode by R5 and R7, the gain is determined by the ratio of R4 to R5 and R6 to R7. C2 and C3 provide AC signal coupling.

IC1e and f are configured as a Schmitt trigger, with R9 providing positive feedback for a sharp switching action. Those noise spikes which pass the switching threshold will produce clean logic level pulses at the output. These random pulses are applied to the clock input of IC4, a flip-flop wired as a divide-by-two circuit whose Q output is taken to the up/down control of IC2. Although the number of random pulses in each IC2 clock period will be pretty much the same, the state of the Q output is arbitrary and will be truly random with an equal chance of a 0 or 1 (or rather, not truly random if one allows that the mind can bias this in some way).

The clock used to drive IC2 is the usual CMOS astable using IC1a and b, with trimming components RV1, C1 and R2. Potentiometer RV1 is a front panel control which sets the frequency of the clock and makes available a large range of rotation speeds.

It was found necessary to limit the supply current with R11, since IC1 (the 4069B) is used in the linear mode as a noise signal amplifier and with too high a supply voltage it can start dissipating enough current to destroy itself!

Fig. 3. Timing diagram of the circuit.
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High-Res Circles
By Anthony DeBoer

THERE'S MORE than one way to skin a cat, as a popular expression goes, and drawing high resolution circles is no exception. The number of ways to draw a circle on a computer's screen, plotter, printer, or whatever device, is probably limited only by the imaginations of the people undertaking to do it. Some methods are much more efficient than others, however, and give better results.

The methods given here, while written in BASIC, are adaptable to almost any language or computer. The last one, in fact, almost begs to be rewritten in assembler. It's the idea that counts, not the coding for a specific machine.

Although the examples are given in Applesoft BASIC, they can be adapted to virtually anything else that can do graphics.

For those unfamiliar with Applesoft, the statement HPLOT X,Y draws a dot at the co-ordinates (X,Y) and the statement HPLOT TO X,Y draws a line from the last point referenced to (X,Y). Each method is presented as a subroutine at line 100 that will draw a circle of radius R centred around the co-ordinates (U,V).

Thus, a simple main program to demonstrate a circle might go like this:

10 U = 70:V = 70:R = 50
20 HGR:HCOLOR = 3
30 GOSUB 100
40 END

Note that the statements on line 20 set up the Apple's high-resolution graphics.

Rooting in the Round
The first method of drawing circles is to use square roots. Many people should remember from highschool math classes that a circle of radius R is defined as the set of points (X,Y) such that R squared equals X squared plus Y squared. To re-express that, if you have a value for X, then Y = ± SQR(R^2-X^2). You could thus, for each X, compute and plot Y. The following program fragment will do this.

100 S = R*R
110 FOR X = 0 TO R
120 Y = SQR(S-X*X)
130 HPLOT U + X,V + Y
140 HPLOT U + X,V-Y
150 HPLOT U-X,V + Y
160 HPLOT U-X,V-Y
170 NEXT X:RETURN

Note that X and Y are computed for one quarter of the circle only, and that then all four quarters are done on each cycle. It will start drawing at the top and bottom, and go all four ways until the curves meet at the sides. The square of R is precomputed as S, in order to save computer time.

Figure 1. Top, the square root circle; middle, one of the sine/cosine circles; bottom, the binary circle. They aren't exactly circular here only because the printer they were printed on wasn't exactly square.

The circle generated by this routine is given in Fig. 1. Note that the curve is densest at the top and thinnest at the sides. This is because the X's are a constant distance apart, while the Y's are not. You could solve this by changing R, in line 100, to R/SQR(2), and adding four more plotting statements, so that you are drawing eight eighths of the circle instead of four quarters. However, there are other ways of solving the problem.

Elementary Trig
Another intuitive way of drawing a circle is to use sines and cosines. If you take X = R*COS(THETA) and Y = R*SIN(THETA), for values of THETA running from zero to twice pi, you get a circle. You could do it something like this:

100 INCR = 2*3.14159/SIDES
110 THETA = 0
120 HPLOT U + R,V
130 FOR N = 1 TO SIDES
140 THETA = THETA + INCR
150 X = R*COS(THETA)
160 Y = R*SIN(THETA)
170 HPLOT TO U + X,V + Y
180 NEXT N:RETURN

In this routine, you are actually drawing a polygon, using straight lines to connect the points. The variable SIDES (which needs to be set by the program that calls this routine) sets the number of sides. Choose 5, and you get a pentagon, and so on. To get a good circle, 40 should be sufficient. Remember that the larger SIDES is, the longer it will take to draw.

Speed of drawing is frequently a major consideration in choosing a drawing routine. If the circles you're drawing are intended to be the eyes of the aliens you're using in your new video game, you want something that will be done in the blink of an eye (sorry about the pun), even if the circles aren't perfectly circular. If, on the other hand, you're doing computer art, you want the best looking circle possible, regardless of how long it may take.

If speed is to be a consideration, bear in mind that not all computer operations
take equally much time. Any operations on integers will be faster than the corresponding operations on real numbers, although integers are not as versatile, especially with trig functions. Addition and subtraction will usually be fastest, with multiplication and division being much slower, and things like trig functions, logs, and exponentiation being the slowest. Thus, \( X^2 \) is usually faster than \( X \), and although the routine just presented could have been done without line 110 by setting \( \text{THETA} = \pi \cdot \text{INCR} \) in line 140 each time around the loop, it is faster the way it is. The repetitive addition, it should be noted, may introduce roundoff error on some machines. If the circle doesn't join itself properly at the right side, you may need to use the multiplication after all. Sines and cosines are the worst offender in this routine, as far as speed goes, however, since they take a lot of computation for the computer to figure out. To calculate them properly, the computer might use what's called a Taylor series, which involves adding together a whole series of terms, each consisting of \( X \) to the \( N \)th power divided by \( N \) factorial. Needless to say, this takes just short of forever. Getting rid of these calculations would be a big improvement, and this can be done.

Quick Trig

There are two ways of doing sine/cosine circles without calculating both every time. One is to start with a point anywhere on the circle and then rotate it around 360 degrees. This can be done as follows:

```basic
100 T = 2\*3.14159/SIDES
110 C = COS(T):S = SIN(T)
120 X = R\*C:Y = R\*S
130 NEXT N:RETURN
```

That's fast, at least for BASIC. The Apple takes a few seconds to do the setup (which it does every time with the first sine/cosine routine), but then blisters around the circles when it's set loose to draw them.

Some computers may, however, already use the table lookup dodge (the Atari is rumoured to, for example). They would have a table built in, containing sine values from zero to ninety degrees, and use it to look up very quickly any sine or cosine value requested. This is great for graphics, where speed is important and accuracy needs to be no more than just reasonable, but is not so wonderful for fancy scientific calculations, where twenty decimal place accuracy is usually desired. With this method, there's a tradeoff between having a huge table with tremendous accuracy and no memory left over for anything else or having a small table and less accuracy.

Binary Circles

The final method we're going to look at is one designed for use in fast machine code routines. It can be used in BASIC too, though, and it produces very nice looking circles.

The routine was developed by J. Mitchener based on methods developed by J. Bresenham. It involves about a page of algebra just to prove it will work and to derive the method, but the idea is that the basic routine will draw one octant (eighth) of a circle one pixel at a time by moving either horizontally or diagonally at each step. The other seven-eights are filled in simultaneously using a mirror technique as described previously. The really attractive feature is that it is done entirely using integers. No sines, cosines, or square roots need be computed, and even the multiplication is by powers of two that can easily be computed using machine-code shifts or by adding the number to itself.

The routine is given here. As before, the variables \( U \) and \( V \) are the centre of the circle, and \( R \) is the radius:

```basic
100 HSTOP U + R,V + R
110 FOR N = 1 TO 40
120 HSTOP TO U + R*C(N),V + R*S(N)
130 NEXT N:RETURN
```

```plaintext
Space Invaders, anyone?
```

Conclusions

The circle routine that you'll actually use depends in part on the machine that you're using. If the machine already has a CIRCLE statement in BASIC, you should be using it already and not reading this article. If your machine can't do sines, or if it can but they aren't accurate enough (or worse, if it can't do real numbers at all) you might have to skip the trig methods. You could always work out the values on a pocket calculator and feed them in as data, though. A pen plotter will need something that goes around the circle once, instead of drawing four or eight pieces at the same time. The last routine could, of course, be modified for this, but the trig routines already work this way. If you need a routine for filled circles (solid instead of outlined, in other words), the first or the last routines given would probably be the best place to start, since they will give you every point on the circle instead of a series of connecting lines. The same applies for computers that don't have line-drawing routines (the good ol' TRS-80, for example), where you need dots instead of lines.

But whatever you do, be creative. There are probably other ways of drawing circles, and likely a number of tricks that you can use on your particular computer to draw the best or the fastest ones possible.
Low Voltage Alarm
R. Sinclair

This circuit was originally used in conjunction with an automatic mains/battery dc supply for transceiver operation. It detected the drop in dc voltage (mains supply normally 13.6 V, battery 12 V) to give visual and audio warnings of the voltage decrease.

It is suitable for any application requiring detection of a drop in normal dc supply voltage with the advantage of an audio warning. This would be particularly suitable in a vehicle where a visual indicator may be easily overlooked.

D1, R1 and R2 provide a stable voltage reference and the preset R2 is adjusted so that D2 lights under normal conditions. D4 is then forward biased, keeping D3 off.

When the input supply causes a voltage drop across Q2, D3 turns off and Q3 now turns on through SW1. The voltage on pin 4 (reset) of the 555 goes positive, enabling the 555 which is connected as an astable. The frequency is varied by R9.

To disable the audio alarm function, SW1 is operated cutting Q3 off, but D3 will still give visual indication.

Capacitor Tester
Andy D’Rozario

During circuit development, situations often arise which call for a capacitor to be checked to ascertain whether or not it is functioning. The circuit shown here gives a quick audio and visual indication of the state of the capacitor by using its most basic property — that of DC blocking.

Half of the NE556 oscillates at approximately 200 kHz and is used to provide a train of positive pulses. These pulses are passed through the device under test to the two LEDs and an audio oscillator formed by the other half of the NE556.

The resultant pulses appearing at point X will vary with the device under test as follows:

- **Capacitor OK** — positive and negative pulse train at point X, both LEDs illuminated, Q1 conducts and an audio tone is produced.
- **Capacitor S/C** — only a positive pulse train at point X, only one LED illuminated, C1 charges up and a lower tone produced.
- **Capacitor O/C** — no pulse train, neither LED illuminated, C2 discharges and the audio oscillator is reset.

A PB2720 piezoelectric transducer was used at the output of the audio oscillator, but a high impedance speaker or earpiece should do just as well.

The tester needs the capacitor under test to be greater than 30nF to produce an audio output reliably, however the LEDs will indicate the correct condition down to 100pF at 5 V supply voltage, and 50pF at 12 V.
ETI Circuits File #2

The first Circuits File (originally published in Summer, 1982) contained huge numbers of modern electronic circuits for the designer and hobbyist and was one of our most successful "Specials" ever. This has naturally led us to publish a second version with the same concept but totally new content. The circuits are not projects but can readily be built into them since, along with the schematic itself, there are notes on the operation and applications of the circuit.

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