

FOR ENGINEERS AND ENGINEERING MANAGERS - WORLDWIDE
 Coser

Solar cell activity quickens as many firms strive to reduce cost and improve efficiency. Solar concentrators are used to boost power output, while computers
are employed to optimize solar arrays. But there is still a way to go before cells can be made by continuous processing. For an overview, turn to page 24.

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## CATALOG SHEET SPECIFICATION COMPARISONS



- Source: CTS Series 201 Data Sheet. Mepco Data Sheet ME1004, Piher Data Sheet F-2002 Rev $7 / 73$
 $01 / 2039633$ - Germany 0711/24 2936 - Italy 02/32 5688 - Netherlands 70/87 4400 - United Kingdom 01/572 6531 Norway $2 / 711872$ • Sweden 764/20 110 • Japan 075/921 9111 • Australia 02/55-0411 03/95-9566 • Israel 7771 15/6/7

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[^0]

## EVER SINCE OUR CREAT MEMORYBREAKTHROUGH THEOLD INDUSTRY STANDARD AINTT WHATIT USEDTO BE.



When we introduced the Fairchild 64K F464, we said it had the stuff to be the industry standard someday. We even started out with a second source. Apparently, a lot of memory people took us seriously. Response has been so positive, we've already increased our production capacity.

Plugging the gap between MOS and magnetic memories, the F464 is the first semiconductor Bulk Storage Devise (BSD).

The F464 is the densest semiconductor memory ever made. A compact die size of less than 40,000 mil $^{2}$ - not much larger than today's 16K RAMs. All packaged neatly in a standard 0.3 -inch 16-pin DIP.

## HICH PERFORMANCE, LOW OVERHEAD.

There has never been a device like the F464.
It's a $65,536 \times 1$-bit dynamic serial memory organized as 16 randomly accessible shift registers of 4096 bits each. The four address bits are decoded on-chip to select which one of these 16 shift registers is to be accessed. Control inputs include Write Enable and Chip Select. It requires standard power supplies of +12 V and $\pm 5 \mathrm{~V}$.
All inputs (except the clocks) are directly TTL compatible. The two high-frequency and two low-frequency clock inputs are low capacitance 12 V signals which can be easily generated with simple logic.

The logic required to generate this 4-phase clock costs less than one memory chip and is generated only once per system.

| Part <br> Number | Maximum <br> Frequency <br> $(\mathrm{MHz})$ | Maximum <br> Active Power <br> Dissipation $(\mathrm{mW})$ |
| :---: | :---: | :---: |
| F464-4 | 2 | 238 |
| F464-3 | 4 | 298 |
| F464-2 | 5 | 336 |

Speed and power dissipation for F464 family.

So you don't pay for memory overhead every time you expand This lower overhead cost means lower system costs to you. Maximum data rates range from 2 MHz (F464-4) to 5 MHz y for the SPS structure is 1 MHz . Since all 16 registers shift simultaneously, the average random access time (called latency) is only $410 \mu \mathrm{~s}$ at 5 MHz - a truly significant performance improvement over other bulk memory technologies! And, at the same time, the power dissipation remains low: typically $3.5 \mu \mathrm{~W} /$ bit at 5 MHz , and $0.6 \mu \mathrm{~W} /$ bit during standby at 1 MHz . Three part types are available to cover a wide range of maximum speed requirements.

These performance benefits make the F464 a natural for hybrid head-per-tracks or fixed-head discs, extended cache, and many other high-density memory applications.

Other outstanding F464 characteristics include solid state ruggedness, speedy data rates and the most semiconductor memory per square centimeter in the industry. Not bad for a small chip! LOW COST, HIGH VERSATILTY.

The new F464 is three to four times less expensive per bit than RAMs. It is also cost-competitive with all fixed-head discs. So there are no trade-offs between price and performance. The F464 gives you the best of both.

F464 applications range from typical computers to atypical portable memories and digital delays.

And once you order, you can rest assured that you'll get prompt delivery. Fairchild has a 250,000-square-foot plant in San Jose, California that's totally dedicated to VLSI technology and production.

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## More-much moreon professionals

Your editorial, "The professionals" (ED No. 15, July 19, 1977, p. 51) agreed with my personal experience. When my wife developed cancer, I learned all I could, and found that many people feel that cancer can be treated as a nutritional-deficiency disease.

Early this year, my wife learned she had a large tumor in the bladder. Shortly afterwards, about $3 / 4$ of the tumor was removed. However, when my wife was released from the hospital, I was informed that the tumor was malignant.

After a checkup about three weeks later, my wife was given a prescription for an antibiotic and told to come back for a second operation. I asked the doctor about vitamin therapy, including laetrile. He ended up saying I could give my wife anything I wanted to as long as I had her back at the hospital for the second operation.

I learned of a doctor who specialized in nutrition. He examined my wife and prescribed a daily consumption of 3 grams of vitamin C, 50,000 units of vitamin A, 800 units of vitamin E, one high-potency vitamin-mineral supplement, 100 grams of pangamic acid and 2 grams of amygdalin (laetrile). WARNING: Vitamins of this potency should never be taken without a doctor's prescription. I obtained those vitamins, and my wife followed her doctor's instructions, including the antibiotic he'd prescribed, until the second operation.

The complete tumor was removed in a short operation, there were no side effects, and I was informed that a second biopsy proved that the tumor was now benign.

After this experience, I feel there is a strong link between nutrition and
cancer. Until proven unnecessary, laetrile should be allowed to be used as part of a nutritional approach to preventing and treating cancer.

Edgar W. Van Winkle
439 Edgewood Place
Rutherford, NJ 07070

I have generally been impressed with the thought, if not the idea, behind most of your editorials. However, that unfortunate epic included in the July 19, 1977, issue, "The professionals," insults not only my own intelligence but Mr. Rostky's.

While vitamins $B_{1}, B_{2}, B_{3}, C$, and $D$ may well have been known to prevent and/or cure many diseases prior to general recommendation by the medical profession, laetrile is known to cure only terminal wealth! Tests with the former substances will cure disease in any test animal commonly used-but laetrile has yet to cause any discernible reaction beyond cyanide poisoning if used in excess. It belongs in the same category as "Dr. Monster's Super Buzz Bomb for Cancer, Arthritis, Ear Wax, and Editorial Ignorance." Neither laetrile nor anything else will cure any of those conditions.

One must wonder whether Mr. Rostky would be so glib if a loved one steadfastly ignored the advice of physician to treat an operable cancer and then died-after thousands of dollars and several years of treatment with laetrile.

To refuse to sanction quackery is true "professionalism."

Richmond E. Young
307 Bakerdale Rd.
Rochester, NY 14616

I hardly ever take the time-that most precious of all commodities-to
(continued on page 12)

[^1]

Giga-Trim ${ }^{\circledR}$ (gigahertz-trimmers) are tiny variable capacitors which provide a beautifully straight forward technique to fine tune $R F$ hybrid circuits and MIC's into proper behavior. They replace time consuming cut-and-try adjustment techniques and trimming by interchange of fixed capacitors.
Applications include impedance matching of GHz transistor circuits, series or shunt "gap-trimming" of microstrips, external tweaking of cavities, and fine tuning of crystal oscillators.


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## Your ten best reasons for buying it are benefits no other tape recorder/reproducer provides

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This fact filled tabloid gives you valuable information about PROM programming technology To get your copy, circle reader service number or contact Data I/O Corporation, PO Box 308, Issaquah, WA 98027
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In a world of claims and counterclaims, one thing is clear. EMM SEMI is still in the lead. Of course, we not only had a healthy head start, but we field a whole family of 4K static RAMs.

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THE TRAVELERS
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CIRCLE NUMBER II

## Across the desk

(continued from page 7)
write emotional letters, but I cannot resist joining the laetrile/professionalism fracas. I have no personal knowledge concerning the effectiveness of laetrile, but I remember when hypnotism was introduced to the public. The AMA immediately attacked the validity of the process and pressed for vigorous prosecution of the practitioners. Some years later when hypnotism was a little better understood, the AMA decided that it was real but could be practiced only by MDs.

Acupuncture has been treated in precisely the same manner. First there was no such thing-now (in California) it can be practiced only by MDs. As Linus Pauling points out in Vitamin C and the Common Cold, the British government knew for years that limes prevent scurvy, but found it more expedient to impress fresh sailors than to protect the health of those on board.
The official reaction to Pauling's book was immediate and predictable. The staff "doctor" of the L.A. Times said in his review that Linus Pauling was not qualified to comment on the efficacy of ascorbic acid because he wasn't even a doctor. What a classic example of the high regard that "doctors" have for themselves.
I remember when the director of the Oak Ridge National Laboratory stated in a press conference that one of the principal reasons for expediting the search for a cancer cure was to eliminate the restrictive emission standards that nuclear power plants are saddled with. He retracted the statement the next day, but the mentality is unchanged. Sickness and medicines are very big business in America.
I'll never have any faith in the medical establishment until it puts human values ahead of profit motives. I appreciate immensely the human values that you express in your editorials and wholeheartedly agree with your sequel editorial in the Sept. 13 issue. If the typical dyed-in-the-wool doctor is a "professional"-much less a healerI'd rather be a ditchdigger than be lumped in the same category as such bigoted charlatans.

Jerry Chamkis
Solar Dynamics, Inc.
3904 Warehouse Row
Austin, TX 78407
(continued on page 204)

# Datel's 16 Bit, Microelectronic D/AConverters 

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- 15ppm/ ${ }^{\circ} \mathrm{C}$ Tempco, max.
- Linearity to $0.003 \%$
- 0 to +10 V or $\pm 5 \mathrm{~V}$ Output
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- Priced from $\$ 119.00^{*}(1-24)$

*U.S.A. domestic prices only


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The machine shop is gone - replaced by a solid-state environment. The separate worlds are coming back together.

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We serve special interests-yours!



Announcing another in a long line of industry first's from ISS—the EFF 735 -the first disk drive of its kind ever to employ an on-board microprocessor.

The advantages of microprocessor power in a disk drive are impressive. Complete internal drive diagnostics. Simplified circuitry because most analog circuits are eliminated. No field adjustments-ever. And a lot more, including microprocessor controlled routines that ease the load on the controller and the mainframe.

The EFF 735 gives you 353.8 megabytes on a single spindle using a fixed and sealed disk. There's one spindle per drive and each drive has its own internal power supply and air filtration system. Average access time is 23 milliseconds.

With our fixed head option, you get another 1.26 megabytes and zero access time.

Besides the microprocessor, the EFF 735

gives you a sweeping lineup of operating and maintenance features. A single phase motor. Dual port capability. A completely electronic tachometer. Total modularity of subassemblies. And truly outstanding serviceability, with no field adjustments and no requirement for special tools -one of the big reasons why your total cost of ownership is exceptionally low with the EFF.

EFF stands for Expandable File Family. The 735 is the first member of this new ISS family, later versions of which will have even greater capacities and capabilities. And all versions will be field upgradable so you can increase performance as your needs increase.

ISS is an operating unit of Sperry Univac bringing technological leadership for the generations ahead. For more details on the new EFF 735, write or call OEM Marketing, ISS, 10435 N. Tantau Avenue, Cupertino, California 95014, telephone (408) 257-6220.

# Speeders beware: The cops can fake you out 

The hide-and-seek played by motorists with devices that pick up police radar a couple of miles before it can identify their speeding cars is about to become "radar poker." The reason? A low-cost decoy transmitter that can be mounted alongside or away from the highway to confound a Fuzzbuster, a Super Snooper or other CB "early warning" systems with false alerts.

The Solure is the first electronic countermeasure that promises to cut down the effectiveness of such cop alerts. Developed by Transportation Safety Associates, the decoy radar operates at 10.525 MHz in the X band, just like a Fuzzbuster and police radar. To be self-sufficient, the decoy is powered by batteries recharged by a bank of solar cells, and operates up to 40 hours on a single charge.

Anticipating countermeasures,

Transportation Safety has incorporated a timer that can deactivate the system when it would be prone to vandalism-from midnight to 7:00 AM.

With a Solure set up every 10 miles or so, a speeder with a Fuzzbuster will not know which is the real radar and which is not.
"We'll be playing poker with him," says Walter W. Friel, director of traffic safety for Washington, the first state to purchase and install many of these devices. "By stationing a patrol car at random in these series, we'll further increase system effectiveness."

Meanwhile, radars are appearing that operate in the K band. The present Fuzzbusters can't detect them and improved models will probably cost twice as much. But for $\$ 30$, says Friel, the X-band decoys can be modified to play speed poker on the $K$ band.

## Ultrasonic transducer tests without contact

Usually, ultrasonic transducers for nondestructive testing have to send energy between the transducers and body being inspected, either through bonding or immersion in a liquid. A new ultrasonic transducer system generates and detects ultrasonic waves simply by being placed close to the surface under test. What's more, the transducer is not confined to examining flat surfaces. It also works in a vacuum and at high temperatures that would destroy bonds and vaporize liquids.

In the ultrasonic transducer, a dc magnetic field interacts with a radio frequency field in the surface of either a conducting object or an insulating material with a conducting coat.

The magnet-rf transducer system operates on a unique phenomenon and does not rely on piezoelectric or magnetostrictive effects. Rf electromagnetic fields are produced in the surface of the
test object by means of planar coils placed adjacent to the surface, according to Dr. Bruce W. Mayfield, who helped invent the transducer system. These coils are energized with 1 to 2 $\mu \mathrm{s}$ bursts of rf at a repetition rate of about 100 Hz . Peak pulse power is on the order of one to a few kW .

A dc magnetic field, which may be produced either by an electromagnet or, in recent designs, by a 3.5 to 4 kilogauss samarium-cobalt permanent magnet, is oriented perpendicular to that of the electric vector of the rf field. The ultrasonic waves are generated in the surface of the material at the rf excitation frequency. The waves are produced on the atomic lattice of the metal under test by interatomic forces generated by rf eddy currents interacting with the magnetic field.

The ultrasonic waves travel through the metal and if a flaw is present acoustic energy is reflected. After an rf burst is applied to the transducer, the coils are switched to a receiver, which amplifies the reflected energy
and displays it as echoes on a scope.
The coil-magnet system generates bulk waves and typically operates at frequencies up to 0.5 MHz . But with proper transducer design there is no practical upper limit, according to Maxfield, currently at Lawrence Livermore Laboratories, Livermore, CA. Normal system bandwidth is one to one-and-a-half octaves.

For substantially higher frequencies, surface acoustic waves (SAW) can be generated, Maxfield notes. In this case, rf bursts are applied to a meander line deposited on a piece of Mylar or other insulating substrate. Once the dc magnetic field is properly oriented surface waves result.

This approach provides a noncritical and noncontact way to produce SAW waves that is simpler than conventional methods requiring edge transducers bonded to the SAW-device durface. Maxfield points out that now SAW waves can be generated in a surface too rough for a conventional SAW-transducer.

One possible way to use this improved SAW system is in the inspection of tubing and pipe now performed by eddy-current testing.

The magnet-rf system is available for license through the Research Corp., 405 Lexington Ave. in New York City, 10017.

## Inverter efficiency goes up with precision Xformers

Precision transformers in a power inverter make it $50 \%$ more efficient by getting rid of power-robbing resistors. A precision input transformer to the transistor power switch guarantees a fixed voltage to each transistor, and a precision output transformer guarantees a fixed output load on each transistor. So resistors for balancing the load are unnecessary.

In a conventional inverter, which is 60 to $70 \%$ efficient, the transformers have a single or center-tapped winding connected to the transistors. Current sharing is set by resistors between the transistors.

In the power inverter from Elgar Corp., San Diego, a de signal is chopped and the resulting ac waveform is fed to a transformer with multiple output windings. Each winding feeds one of many transistors in parallel, a configuration chosen to share the inverter's current load and increase poweroutput. The transistor outputs are fed
into another transformer, which has multiple input windings. This transformer raises the voltage to the desired level, while isolating the transistor circuitry from the load.

Without resistors, efficiency climbs to $90 \%$, according to John Waterman, Elgar marketing vice-president.
This efficient inverter circuitry is used in a series of Elgar power inverters designed for harsh environments. But in a commercial package the Elgar inverters will cost about the same as older designs, says Waterman.

CIRCLE NO. 318

## Fiber-optic connector cuts interface losses

Connecting the ends of individual fiber-optic strands instead of bundles cuts the interface losses of a fiber-optic connector from a typical 3 dB to 1 dB or less.
Both electrical and optical lines are housed in this connector, which has been developed by Hughes Aircraft's Connecting Devices Div. for airborne applications.
Dead space between round fibers in a bundle prevents optical conductors from being precisely aligned. Losses usually hover around 3 dB . That is, fully half the light is lost in passing through an optical interface between bundles. Previous loss-reduction techniques have included arranging the bundle in geometric shapes, but the actual fiber-to-fiber alignment was left pretty much to chance.
A single optical fiber is all that's necessary to transmit data from one place to another, and by eliminating all redundant fibers Hughes has developed an optical interconnect scheme that reduces the interface loss to 1 dB or less.
"Axial alignment is the most crucial consideration in an optical connector," explains Jim Wittmann, Military Products Manager for the Hughes division in Irvine, CA. "A coaxial misalignment of a mere 500 microinches will create an optical loss approaching 1 dB ." Angular misalignment and a nonoptimum gap between the two ends add to the interface loss-a .005 in. gap causes a $2-\mathrm{dB}$ loss. But if two ends actually meet, shock and vibration will destroy the smooth glass faces. The ideal gap, according to Wittmann, is .001 in., which causes a minimum loss of 0.5 dB .
It is possible to reduce the alignment losses to zero. So far, Hughes has
demonstrated an over-all loss as low as that by crimping soft bushings over the ends of glass fibers. Machined bushings provide mechanical stability so the fibers can be securely mounted in a two-section alignment tube, one section on either side of the interfacial gap. The gap is controlled by a 1 -milthick washer. Mechanical springs bias the bushings together against the washer, so the assembly stands up to shock and vibration.
Three moisture-proof seals of elastic silicone keep atmospheric contaminants out of an interface, which is recessed $1 / 4 \mathrm{in}$. inside the connector body.

The new optical connector is designed to support a multiterminal data bus for carrying avionics control signals in , weight-critical military aircraft. It meets MIL-C-85028 in lieu of a MIL Spec on airborne optical connectors, according to Wittmann. The units go for $\$ 2$ to $\$ 3$ per mated line.

CIRCLE NO. 319

## Brain cells monitored by tiny FM transmitter

A miniature FM transmitter may help shed light on the workings of that most complex and mysterious of all organs, the brain. A $16-\mathrm{by}-16-\mathrm{mm}$ cylinder attached to a tiny electrode, has been inserted into a laboratory animal's brain to monitor the electrical signals of a single cell. The animal can go about its business, unimpeded by the long and cumbersome wires that were necessary in past studies.
The transmitter was developed by David Pettijohn, a research associate and electronics specialist in the Dept. Psychology at the Massachusetts Institute of Technology in collaboration with Dr. Howard Eichenbaum, a postdoctoral fellow and Anne Deluca, a research associate.
This brain-cell instrument, according to Pettijohn, may revolutionize the rapidly expanding field of psychology, which records and studies the firings of single brain cells in response to specific stimuli. The long-range goal, he notes, is to better understand how information flows within the brain.
Individual brain cells range from 10 to 50 microns, and recording from them requires a tiny high-impedance electrode flexible enough to move "as the brain sloshes around."
The electrode is composed of a bundle of platinum-iridium wires, each 25 microns in diameter. To be stiff enough
to penetrate the brain, the bundle is encased in a shaft of dextrose, which provides temporary support until it dissolves in the brain. The individual wires then can spread out and move with the brain fluid.
Weighing only fourgrams, the transmitter is manufactured by Midguard Electronics, Newton, MA, under license to MIT. The unit contains a FET and four other transistors to amplify and broadcast the signal from the electrode. The FET provides high impedance at the transmitter input and drives a high-power-gain Darlington amplifier whose output in turn provides the signal to the output oscillator.
Powered by a standard $1.5-\mathrm{V}$ hearing-aid hattery, the transmitter has a broadcast range of 10 meters (about 30 feet). Bandwidth is 1.5 to 12 kHz and is pretuned durng assembly to a broadcast frequency specified by the user.
Currently, single cells can be recorded from several days to two weeks. About 30 of the transmitters have been manufactured to date and are being tested in many laboratories.
In addition to animal studies, the transmitter may prove useful for recording electroencephalographic data for studying epilepsy, exercise physiology and neurological diseases.

## $\mu \mathrm{p}$-controller pen <br> gauges signatures

A piezoelectric sensor pen and tablet can, under microprocessor control, analyze signatures by measuring not only the pressure exerted by the pen on paper, but also the pen's accelerations. Developed at Sandia Laboratories, Albuquerque, NM, the pen was described at the International Electron Devices Meeting in Washington, DC, earlier this month.
The pen has a central, slightly flexible shaft that supports two long, thin sensors mounted $90^{\circ}$ apart. The shaft is mounted within a rigid tube; one end is fixed to the tube and the other has a ballpoint-pen tip. When the pen is in use, the shaft can flex, and the accelerations of the pen can be measured in $x$ and $y$ coordinates. A sensor platen detects the third axis of informationpen pressure.
The pen's measurements are more valuable than Fourier analysis of pressure outputs or voiceprints in applications such as computer-room security, says Errol Eernisse, Supervisor of the Soldi State Device Physics Division at Sandia.

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AMP has a better way.
 ordinary contacts is shown by these electron images.

# Solar-cell technology 

Little by little, solar-cell technology is being directed by the federal government toward increasing the nation's energy supplies. While solar-electric, or photovoltaic power systems have a long way to go before they actually compete with fossil-fueled power plants, there are signs indicating that progress is being made:

- Commercial use of photovoltaic power in the field has increased to 500 kW , from 100 kW in 1970.
- The 1978 price of a unit of solar power, a peak watt ( $\mathrm{W}_{\mathrm{p}}$ ), is expected to get down near $\$ 13$-from $\$ 90$ in 1970.
- Silicon-fabrication techniques are breaking away from the 20 -year-old Czochralski growth process and edging into high-volume and potentially lowcost silicon ribbons and sheets.
- A dozen materials are being studied for their photovoltaic effect; singlecrystal silicon still dominates, but polycrystal silicon, gallium arsenide and cadmium sulphide are getting increased attention for special uses.
- Optical concentrators are increasing the effective area of solar arrays by converting more incident energy than would otherwise fall on the cells.
- Computers are helping to configure and optimize complete commercial photovoltaic power systems, according to load, geography, weather and even the cost of money.

The impetus is free energy-a cloudless summer day with the sun shining brightly yields a peak energy distribution of about $1 \mathrm{~kW} / \mathrm{m}^{2}$, the maximum solar irradiation defined by a peak watt. Just lying in the sunlight, a $2-$ in. diameter solar cell provides $1 / 4$ $\mathrm{W}_{\mathrm{p}}$, a 3 -in. wafer $1 / 2 \mathrm{~W}_{\mathrm{p}}$, and a 4 -in. wafer, $1 \mathrm{~W}_{\mathrm{p}}$.

Each unit cell outputs a fixed volt-

## Dick Hackmeister

Western Editor


Czochralski-grown silicon crystals with up to a $5-\mathrm{in}$. diameter are sliced waferthin to produce solar cells. After diffusion, the discs are metalized with an electron-collecting grid. It becomes one of the output terminals. (Motorola).
age, about a half-volt (see box). The cells are normally configured in seriesparallel arrays to provide a specific voltage and current capability. Except for the cost involved, there is no limit to the number of cells that can be interconnected.

## The price volume quandary

But solar-cell technology faces a dilemma. Without high-volume orders,
meaningful price reductions can't be realized; without low prices, high volume orders can't be placed. And the 1977 price for a peak watt, $\$ 15$, is three times what it should be if photovoltaic power is to enjoy widespread use.
For three years the Energy Resources Development Administration (ERDA), now part of the newly created Dept. of Energy, has been funding OEM manufacturers and universities in an over-all solar-electric cost-reduc-

## advances-but slowly

tion program, which relies on the federal government's three existing research facilities. The Jet Propulsion Laboratory in Pasadena, CA, researches and develops raw solar-cell material, fabricating large-area sheets, encapsulating the finished cells, integrating them into modules and testing the completed array. Sandia Laboratories in Albuquerque, NM, designs and evaluates optical concentrators and "total energy" systems, and power-conditioning and storage subsystems, and identifies over-all system tradeoffs. Lewis Research Center in Miami, OH , tests and demonstrates real-life systems in the field. Through Lewis, the largest solar-electric array was installed and set into operationa $25-\mathrm{kW}\left(\mathrm{W}_{\mathrm{p}}\right)$ irrigation-water pumping station in Nebraska.

## Introducing SERI

Soon, however, federal solar-energy R\&D will be consolidated under the roof of the Solar Energy Research Institute (SERI) in Golden, CO. Created by Congress, SERI is just now getting staffed and organized under the directorship of Dr. Paul Rappaport, recognized as one of the country's leading experts in photovoltaic conversion.

Since petroleum is the most portable energy source around, Rappaport sees its use only for the transportation industry. Stationary facilities will ultimately be powered by the sun, Rappaport predicts. But more efficient solar-cell manufacturing processes are needed before that can happen. Because low-production volume requires a great deal of manual intervention, Rappaport advocates shifting away from the batch-oriented schemes to a continuous-flow process that can be automated. "Just sawing a boule of silicon into wafers adds 35 cents to the price of each watt," he notes.

Most of today's commercially avail-


Even the best solar cells won't reach more than $21 \%$ conversion efficiency, according to theory. Today's devices range from 8 to $18 \%$.


Generating $1000 \mathrm{~W}_{\mathrm{p}}$, this two-axis tracking array uses fresnel lenses to concentrate solar energy 60 times. Lenses are $80 \%$ efficient: grooves are turned inward to prevent dirt from accumulating.


This map shows the total daily energy in Langleys falling on the U.S., in the course of a year ( $1 \mathrm{~L}=1.162 \mathrm{mWh} / \mathrm{cm}^{2}$ ).
able solar cells are made by the same process that produces their cousins, discrete transistors and integrated circuits. Round wafers are made by growing a cylindrical ingot, or boule, of high-purity, single-crystal silicon. The Czochralski process, which grows the silicon boule from a small silicon crystal, makes wafers up to 5 in . in diameter (see photo). Slicing the boule like a salami leaves many discs standing on edge. They are then cut away from the common spine. Separated from the spine, each disc of silicon is made into a single, large diode (see box) by the same diffusion process that creates ICs and transistors, except that the wafer surface isn't masked and only a single diffusion step is necessary. (Interestingly, it makes no difference whether the $p$ side or the $n$ side of the junction faces the sun.)

Finally, an ohmic, current-collecting grid is deposited over the face of the wafer. It becomes one of the solar cell's output terminals-the other terminal is formed by metal deposited on the backside of the cell. Output voltage polarity is determined by the direction


Solar cells are arranged in commercially available modules to fit the application. These units, which are available from Solar Power Corp., North Billerica, MA, generate $25 \mathrm{~W}_{\mathrm{p}}$. $9.2 \mathrm{~W}_{\mathrm{p}}$ and $1.4 \mathrm{~W}_{\mathrm{p}}$, respectively.
of the p-n junction; current flows in the forward direction.

## The ultimate in efficiency

Growing silicon in boules is great for producing quarter-inch IC chips (the wafers are scribed, then cracked into squares). But the process is woefully inappropriate for making carpet-sized


The largest photovoltaic installation yet is this $25-\mathrm{kW}_{\mathrm{p}}$ array at an irrigationwater pumping station in Nebraska.
sheets of the material to collect and convert sunlight. Several firms are currently under contract to the Jet Propulsion Laboratory to produce large-area silicon, either in ribbons or in sheets. These companies include Mobil-Tyco (Waltham, MA), Motorola (Phoenix, AZ), RCA (Princeton, NJ), Rockwell International (Anaheim, CA), Honeywell (Bloomington, MN), Westinghouse and General Electric (Schenectady, NY).

Other JPL contractors are working to alter the existing batch-manufacturing process with such alternatives as ingot casting and improved sawing or cutting of the boules. These contractors include Texas Instruments (Dallas, TX), Varian (Lexington, MA), Coors (Golden, CO), Crystal Systems (Salem, MA) and Eagle Picher (Miami, OK).

Meanwhile, commercially available solar arrays are beset with problems. For one thing, they are only about $12 \%$ efficient-that is, they convert incident sunlight into usable electric power at the rate of $120 \mathrm{~W} / \mathrm{m}^{2}$. Small-area laboratory devices have reached efficiencies as high as $18 \%$, but theoretically no cell can exceed $21 \%$.
For silicon, unmatched spectral response reduces the available photons to $44 \%$ efficiency. The band-gap limit further reduces them to $29 \%$, and fill factor brings the number down to $27 \%$. After surface reflections and defects, the figure is $21 \%$, which still doesn't allow for dirt, metal-mask shadows and $I^{2} R$ losses.
Geographical location has a profound effect on a solar array's output. Peak watts are generated only during the ideal conditions of summer sun and clear skies. Even then, peak watts occur only at noon. Averaging peak power with cloud cover, night darkness, and the sun's acute angle of incidence in winter cuts the annual amount of usable power down to $25 \%$ of the peak in Phoenix, and down to $16 \%$ of the peak in Boston (see map).

Temperature takes a toll, too, and a solar array with a supposed $15-\mathrm{V}$ output will droop to 13 V if noontime temperatures hit 110 F ( 43 C). Evaluation arrays purchased by JPL are spec'd for operation at 150 F ( 65 C ), but that's probably overkill for most OEM applications.

## Problems on earth

Properly interconnected and potted, solar cells can be expected to operate indefinitely. In fact, solar cells have

been used in space for years, and the Jet Propulsion Lab has not been able to find a measurable degradation or failure yet. But back on earth, a few reliability problems do crop up. The chief culprit, not surprisingly, is the encapsulating, or potting material of a solar-cell module.
Although a module needn't be hermetically sealed, it must be weatherproof and impervious to moisture so that it can withstand day/night temperature cycling. Moisture on the cell's surface degrades the unit's performance rapidly. It causes extra losses by absorption and refraction, and it promotes the growth of organic substances that block valuable sunlight.

Another performance degrader, oddly enough, is the sun, whose ultraviolet rays darken the encapsulant material.
There is even a problem with the most used encapsulant, silicone rubber. Its sticky surface collects dirt. Manufacturers cover the module's face with clear plastic or tempered glass to make the module better able to clean itself: The surface should wash clean in the rain, and snow should slide off.

Cell interconnections within the module can also become failure mechanisms, so multiple, redundant connections must be made to the disc.

Since solar arrays make power while the sun shines but never at night, power conditioning and storage facilities are prime requisites for any commercial solar-electric system.
Automobile-type, lead-acid storage batteries remain the only realistic power buffers. Vendors combine them with solar-cell modules, a shunt regulator, a blocking diode (to prevent shorting out the batteries at night), mounting hardware and cabling to configure a complete OEM solar power generator.

System configuration is called "sizing," which normally includes a computer analysis of the site's latitude, longitude, altitude, mean temp and yearly "in-sol-ation"-which refers to the amount of time the site is in the sun, or "in sol." The unit of measure of insolation is the Langley, which is equal to $1.162 \mathrm{mWh} / \mathrm{cm}^{2}$ (see map).

Armed with a description of the power required and the site's solar circumstances, a computer determines the most efficient system for that application: the number of solar modules and their series-parallel connections, the array's compass heading and tilt angle, the number of batteries required and their interconnections, as well as projected system performance on a


This two-axis, three-section tracking concentrator generates $300 \mathrm{~W}_{\mathrm{p}}$, and uses a solar-powered $\mu \mathrm{P}$ and stepping motors to follow the sun. RCA expects this "fly's eye" concentrator to cost $\$ 1 / W_{p}$ in volume quantities.


A nontracking, solar concentrator uses high-density strips of cells and parabolic reflectors to generate 100 $W_{p}$. (Argonne National Laboratory).
monthly basis.
Computers can drive the price of flat-plate solar arrays as low as process and materials will allow, but concentrators can accomplish optically what process and materials can't-an effectively large solar cell. Concentrators are considered the "hot" price-reduction strategy, and are being intensively developed; some may become commercially available as early as mid-1978, after testing at Sandia.

## Concentrating on the sun

Concentrators can gather and focus more of the sun's energy on available solar cells using parabolas, fresnel and optical lenses. Tests at Sandia show that concentrators can boost the equivalent efficiency of silicon solar cells above $15 \%$, and gallium arsenide cells to almost $20 \%$; they are most effective in the southwestern United States where sunlight is most intense and least diffuse. Both tracking and nontracking concentrators show promise.

Nontracking concentrators effectively multiply the surface area of a

## The 'Un-led,' and how it works

Operating, the solar cell is a lightemitting diode working backwards.

A conventional LED, like the type that might be used as a display in your watch, has a depletion layer formed in the semiconductor material by oppositely doped impurities. An external energy source-voltage -excites the electrons to conduct through the depletion layer. Excess energy is given off as photons, which you can see.

A solar cell, too, has a depletion layer formed in the semiconductor material by oppositely doped impurities (the cell is a p-n junction). But it works in reverse. Photons excite the electrons, and the junction develops a voltage.

Working on different sides of the street, a LED doesn't emit a broad range of wavelengths, and a solar cell doesn't absorb a broad rangewhich accounts for the cell's losses due to imperfect spectral response.

The solar cell and LED do share limitations. The energy band gap ( $\Delta$ $\cong 1.5 \mathrm{eV}$ ) limits the amount of energy either one can convert. And because they're less-than-ideal diodes, both the LED and the solar cell suffer reflection losses as well as losses due to electron-hole pairs recombined in impure areas.
solar cell as much as 20 times ( 20 suns). They use parabolic shapes to collect extra photons and form them into a beam, like that of a flashlight. An array of concentrators and a like number of solar cells are mounted on a rack that looks like a backyard swing. The rack tilts the array at the best angle for the site's latitude, and is adjusted seasonally.

Beyond a concentration ratio of 20 suns or so, the optical alignment of san, lens and cells becomes too critical for a fixed-frame array-because of the earth's rotation. The sun must be tracked.

The tracking speeds involved are slow enough for stepping motors to move the frame of a tracking concen-trator-it's all under $\mu \mathrm{P}$ control. Together, the stepping motors and the $\mu \mathrm{P}$ consume only a small fraction of the energy they help collect.

The highest concentration ratio to date is in RCA's "fly's eye" (see photo). With a concentration ratio of 300 suns, the tracking must be very accurate: the unit updates its position 30 times a minute...

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# IC packages are changing shape to handle growing LSI chips 

Pressure is mounting for the redesign of IC packages, both to accommodate the increased size and complexity of LSI chips and to bring down the cost of consumer ICs. As a result, new IC-package standards have been proposed and new packages are being developed.
Putting more and more circuitry on a digital chip is not only increasing chip size, but in effect is making the standard DIPs too small. But lengthening the packages to add more pin-out connections extends internal leads and adds parasitic capacitance that loads down the circuits, limits operating speeds and adds undesirable delays.
Moreover, the rectangular shape of the DIP packages requires excessive board area. This holds down the number of packages that can be mounted in a given space.

Meanwhile, high-volume, low-cost consumer applications such as calculators, clocks and video games, are suffering because the IC package now costs as much or more than the chip.

The standard DIP cavity, 300 mils wide, is just getting too small for the new memories being developed, says

## Jim McDermott

Eastern Editor

John Hewkin, strategy manager of MOS Memories for Texas Instruments in Houston. For a memory that will take only 16,18 or 20 pins, the chip is getting to be almost too big to fit into a standard package.
"For example, our new 64-k RAM will barely fit into a 16 -pin, $300-\mathrm{mil}$ package," Hewkin observes. "We could live with a 400 -mil package width for LSI for the foreseeable future, but the $300-\mathrm{mil}$ unit is marginal."
The industry has acclimated itself to the specific DIP sizes because of the massive investment in packaging equipment. But agreement is almost universal that new package designs are needed. Hewkins points out that TI has tentatively put off making wide memories because of existing package limits.
However, in what has been termed a rare show of collaboration, new standards for LSI packages are being coordinated by package users, package manufacturers, and the semiconductor manufacturers.

Tentative specifications have been developed by the Joint Electronic Device Engineering Councils for a family of LSI chip-carrier packages that will permit a system user to change from one package to another without redesigning the system.

Two square-package series-one for

PC-board use, the other for high-density ceramic packaging-are being proposed by the JEDEC JC-11.3.1 Task Group, headed by Daniel I. Amey of Sperry Univac, which spearheaded the industrywide effort over two years ago. Mechanical details and dimensions have been resolved. According to Amey, pin-identification details still have to be finalized.

## Square is fine

A square format was determined to be most efficient in terms of circuit density. Also, the format minimizes the lead length of all connections, and consequently reduces parasitic lead capacities. The basic package has contact pads for interfacing with a connector or mounting directly to the PC board or ceramic substrate. The leads on the leaded version are located on the sides to act as spring contacts for mating with a socket.

Type I, geared to PC-board use, has contact pads or leads located on $50-\mathrm{mil}$ centers. Designed for $28,44,52,68,84$, 100,124 and 156-lead packages. Type I ranges from about 0.45 in . to 2.05 in . square.

The Type II series proposed by JEDEC is like the Type I series except that the leads are on $40-\mathrm{mil}$ centers,


New LSI IC packages proposed by JEDEC have features shown above. One style has leads on 50 -mil centers and
is for PC-board mounting. The other style, with leads on 40 -mil centers, mates with ceramic substrates.


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A 68-lead IC-chip carrier, lower center, has been developed by 3-M to be one of the new 50 -mil centerline LSI packages being standardized by JEDEC. Chip carriers are also shown.
rather than on 50. Primarily designed for high-density ceramic packaging with direct attachment to ceramic substrates, Type II devices would have 40 , $48,64,80$ and 96 pins.
Both leadless and leaded variations of Types I and II have been proposed (see figure). The design of these new leadless and leaded types will affect mounting requirements. For example, leadless type $A$ is intended to be mounted with the chip bonded to the top surface for better heat dissipation. A mating connector is required.

Leadless type $B$ can be mounted directly on a ceramic substrate with the chip on the bottom of the carrier. For PC-board mounting, however, a connector is needed.

Leaded type A, a molded plastic nonhermetic package, is designed for reflow soldering to either ceramic or PC board substrates. It is intended for continuous-carrier-film automated die attachment and package mounting. Leaded type B is simply a leaded version of leadless types A and B, aimed at reflow soldering.
Even before JEDEC came up with proposed DIP standards, companies were responding to the packaging problem. For example, a 28 -pin leadless design by General Instrument called the Mini-Pak eliminates the lead frame, molds and molding compounds, which provides substantial savings in consumer applications. Interestingly enough, the Mini-Pak is now a part of
the proposed JEDEC standard.
The Mini-Pak board is made of an epoxy or composite laminate on which copper metallization is produced. The copper pattern is first nickel-plated, then gold-plated for wire bonding.
The LSI chip is bonded to the laminate both electrically and mechanically with epoxy adhesive, and connected to the 28 termination pads with aluminum wires. The completed MiniPak is encapsulated and dipped in solder to coat the termination pads.

Anticipating the switch over to JEDEC's proposed standards, both the Electronic Production Div. of 3-M, St. Paul, MN, and Kyocera International in Cupertino, CA, are tooling up for a 68 -pin 50 -mil-center-line ceramic version of the JEDEC package. Fairchild is expected to use this for high-speed ECL logic. This particular chip carrier is about $0.95-\mathrm{in}$. square and has 68 contact pads to mate with a connector.
Several firms are developing prototypes of connectors to be used with the proposed JEDEC designs. These include Berg Electronics, Cumberland, PA, Texas Instruments, Attleboro, MA, and AMP Inc., Harrisburg, PA.
Interestingly, AMP had decided months ago that a new low-cost chip package was needed. So it went ahead before the JEDEC proposals were finalized and developed a flat, 24-pin package for mass application. The AMP package, called a "premolded packet," also has its leads on $50-\mathrm{mil}$ centers, six to one side. The package is slightly less than $0.4-\mathrm{in}$. square.

The AMP device can be soldered easily to PC boards, even those that may warp during soldering operations. The packet frame is molded with six leads extending from each side. The chip is inserted inside the carrier, bonded, and capped after being filled with silicon gel for chip protection.
The leads are folded back underneath the flat pack for a springlike effect. As the PC board expands and contracts during a soldering operation, the compliant leads protect the package itself from any strain. On some boards, for example, ceramic carriers can't be soldered because PC-board strains produced by soldering can crack the fragile carriers.

Industry concensus is that the standard 14, 16, 18 and 24-pin DIPs will be around for a long time yet in many lowcost, noncritical applications. For the LSI circuitry, however, the new 40-pin and up packages will be needed..■

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# Washington report 

## Another \$1.5-billion slated for B-1

The Air Force will spend another $\$ 1.5$-billion to test the B-1 bomber as a possible alternative to cruise missiles, despite President Carter's decision in June not to order the bomber into production.
The figure came to light when the Air Force submitted to Congress its quarterly Selected Acquisition Report (SAR), which lists the projected total costs of the Pentagon's 45 largest weapons-systems programs. The latest SAR estimates that B-1 costs will total $\$ 4.9$-billion, or $\$ 19.9$-billion below the previous estimate, which was projected before the decision not to produce 240 aircraft. The Air Force has already spent nearly $\$ 3.5$-billion on three prototype B-1s and will get a fourth in early 1979. That aircraft will be used to test its special defensive electronic countermeasures system, and the Air Force expects B-1 flight tests to continue for at least another three years.
The latest SAR projects that all told the Air Force's top 45 programs will cost $\$ 179.5$-billion, or $\$ 19$-billion less than projected in the last report. Program increases will be worth $\$ 900$-million, with the largest to be an extra $\$ 103$-million for 12 Navy A-7E attack aircraft, which were added by Congress to a program that the Pentagon has been trying to end for at least three years.

## Air Force sensor to detect aircraft from space

An experimental sensor system scheduled to be launched into orbit by the Air Force in March 1981 will be used to detect enemy strategic aircraft from space. The first sensor package, whose code name is Teal Ruby, is being built by Rockwell International Corp.'s Space Div. at Downey, CA.
If it becomes operational, the system will provide early warning of an attack by the Soviet Union's new Backfire long-range supersonic bombers and, possibly, its cruise missiles. Operating frequencies are classified, but it is believed that the Teal Ruby sensors will use the infrared spectrum to detect the effluent gases of the bomber and cruise-missile engines. Each type of engine has its own identifiable spectral "signature," which can be used to locate the engine's aircraft.
Rockwell has completed the design of the Teal Ruby sensor and has recently received a $\$ 3.3$-million contract from the Air Force's Space and Missile Systems Organization to build the first sensor package, which will be launched from NASA's Space Shuttle. The program has cost $\$ 24$-million to date.

## Battlefield sensors are going automatic

Prototypes of an automated battlefield sensor system to detect movements of enemy troops are being built by RCA for testing by the Army Electronics Command (ECOM). The sensors in the Remotely Monitored Battlefield Sensor System (REMPASS) will work like the unattended ground sensors used in Vietnam except that they will use a more advanced data-communications system.

The REMPASS's magnetic, seismic, acoustic and infrared sensing devices remain passive until activated by enemy movements. Then they send back signals via vhf radio link, directly or by means of repeaters either on the ground or in helicopters.
At least six engineering models will be delivered to ECOM. RCA's Automated Systems Div., Burlington, MA, recently won a $\$ 9$-million contract from ECOM to integrate the sensors with the improved communications system.

## Power system for space is on the way

An isotope power supply capable of generating up to 2 kW for spacecraft is being developed for the Department of Energy. The Brayton Isotope Power System (BIPS) is a closed-cycle gas-turbine electrical generator that takes the heat generated from a plutonium isotope and converts it into electric power through a generator driven by a high-speed turbine. Prime contractor Garrett AiResearch Manufacturing Co. of Phoenix is testing it in an altitude chamber simulating space conditions.

The generator is due to complete a 1000 -hour endurance run next May. If it passes, a BIPS may be tested on NASA's Space Shuttle in late 1981 or early 1982. The ultimate goal is 50,000 hours, or nearly six years of operation. An earlier $15-\mathrm{kW}$ Brayton power system developed by Garrett for NASA recently passed the 30,000 -hour endurance mark at the agency's Lewis Research Center in Cleveland.

## Superconductive receiver operates in millimeter region

The first successful operation of a superconductive heterodyne receiver in the millimeter-wave region has been reported by scientists at the National Bureau of Standards Cryogenics Div., Boulder, CO. Superconductive methods have already been demonstrated for microwave receivers and magnetometers.
The millimeter-wave receiver is tunable over the 200 to $325-\mathrm{GHz}$ waveguide band. The system's noise temperature (single-sideband) is estimated at 823 K , with an instantaneous bandwidth of 20 MHz . This temperature is an order-ofmagnitude improvement over the best uncooled mixer receivers at this frequency, according to NBS, and is comparable to the noise performance of helium-cooled indium-antimony bolometers, which are limited to i-f bandwidths of less than 2 MHz .

The mixer is a point-contact Josephson junction mounted in a single-mode wavguide. A $300-\mathrm{GHz}$ backward-wave oscillator is used as the local oscillator to produce a nominal $9-\mathrm{GHz}$ i-f signal, which is amplified by a low-noise ruby maser.
One potential application for the new receiver is astrophysical observation of the line spectra of molecules.

Capital Capsules: The Office of Federal Procurement Policy is expected to reaffirm the government policy of contracting out as much work as possible to industry rather than having it done "in-house" by government employees. This policy is required by Office of Management and Budget circular A-76, but inflation and tight federal budgets have forced government agencies to keep work in-house in order to protect employment levels. The new policy statement is expected to require Commerce Department certification that no commercial source is available before in-house activity can be undertaken....NASA is reviewing all its field centers work forces to meet President Carter's requirement that its total work force be cut back by $\mathbf{5 0 0}$ jobs this year. Sources say NASA may have to close a field center.


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## Editorial

## I hear you

A story is making the rounds about this 70 -yearold gentleman whose doctor urged him not to marry his 25 -year-old ladyfriend. "It could prove fatal," the doctor warned.
Well, the old gent didn't have complete faith in his doctor. He saw him mainly as a specialist in extracting fees. But he was troubled. So he stroked his chin for a few minutes as he meditated. Finally the old campaigner arrived at a decision: "I'll take my chances. If she dies, she dies."
It's likely that this tale is apocryphal. But it's not likely that our hero was the first-or last-to hear just what he wanted to hear. The world is full
 of people with one-track ears.
Look at Charlie, the chief engineer at a small test-equipment company. When reminded of his large staff turnover, and the extraordinary number of his engineers who left to work for his competition, Charlie saw this as a tribute -proof that the training he provided made superior engineers.
If you were to tell Charlie that some of his working conditions made morale so poor that he was actually spending lots of money to train staff for his competition, he wouldn't hear it. He couldn't hear it. He could hear only praise.
Jack, on the other hand, could hear only condemnation. If you were to tell Jack that his new design was splendid, but that it might be improved with a slightly larger display, he would arch up and claw at you. Even the mildest suggestion was an attack.
Me? I'm different. I hear exactly what people tell me. There's never any coloration due to my background and make-up. How about you?


George Rostiy
Editor-in-Chief

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Electronic Design's popularity overseas continues to increase - especially in Europe. The growing fraternity of Electronic Design subscribers is now pushing toward the 100,000 mark - and more than 16,000 are located outside the U.S.A. In fact, we now have more EOEM engineering managers and engineer subscribers in Europe than any

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other U.S. journal. Engineers throughout the world look to the U.S. for leadership, and they turn to Electronic Design for the same reasons you do. If your company sells in Europe, or is thinking of starting up, pass the word to your marketing people. Maybe they aren't aware of our power, influence and authority among key specifiers overseas.


You may have heard our name being bandied about when it comes to coils, transformers and filters. These are just three of the secondary products we've been working on for years in combination with our ferrite core material and ceramic capacitors. But have you heard that these items are just a small part of our total product repertoire?

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thermistors (heat-sensitive elements), electric wave absorbers and many more. We supply them to the world's assembly makers.

With our ferrite cores for communications equipment we combine proprietary low-loss and high-permeability materials with sophisticated processing technology. Then again, we're kept busy supplying the world's music fans with cassette tapes - virtually all audiophiles have heard about our ED and SA series. Even our tapes, however,
take their cue from our magnetic material technology which began with our ferrites.

We already hear whispers about us in industries across the world like automobiles, business machines, medical electronics and industrial equipment, which are starting to use our parts. Although we're sure that every thing you hear about us is true, we only hope that you don't mistake the part for the whole.

## Technology

# Drive servos with a switching amplifier. <br> With a switching servo amplifier in your loop, you get high efficiency in a compact control package. 

Delivering power with low losses in the output stage, a switching servo amplifier (SSA) requires far less heat sinking than a similarly rated linear amplifier. The switching mode power output provides typically $\pm 75 \mathrm{~V}$ at $\pm 10 \mathrm{~A}$ to drive a dc servo motor. And whether you're building a position or velocity servo, an SSA controls equally well in either case. Of course, you can build your own servo amp stage by stage, but an SSA module allows you to eliminate hardware design of your control unit.

Within an SSA package you'll find all the conventional servo amplifier stages-summing block, gain stage, compensation network and power amplifier. System control is performed by this single moduleyou supply a motor and feedback elements to complete the servo loop.

For position servos, the feedback element can be a potentiometer, resolver, or any element that provides electrical output as a function of position. Tachometers are often used as feedback elements to sense velocity. An SSA's input summing block simultaneously accepts signals supplied by two feedback elements and a reference source.

At an SSA's output, power is supplied to a motor with high efficiency. Dissipation in the amplifier is minimized by switching mode operation, with the output transistors connected in a bridge circuit.

## Crossing the " H " bridge

In an SSA's power amplifier, four output transistors are operated in the switching mode, in an " H " or "transistor bridge" configuration.
In the SSA block diagram of Fig. 1, GB represents the power stage, including the control electronics. The " H " bridge portion of $\mathrm{G}_{\mathrm{B}}$ is shown schematically in Fig. 2. Arranging the output transistors, $Q_{1}$ through $Q_{4}$, in a bridge circuit allows current to flow in two directions through the motor, which consequently develops torque in both the clockwise and counterclockwise directions.

When $Q_{1}$ and $Q_{2}$ are turned on by a driver, current flows in the direction of the current arrow, $i(t)$, in Fig.

Dana F. Geiger, Chief Engineer, PMI Motors Div. of Kollmorgen Corp., 5 Aerial Way, Syosset, NY 11791.


1. A complete servo amplifier is housed in a single switching servo amplifier (SSA) package. Depending on feedback elements, either a position or velocity servo can be built. In the SSA block, $G_{B}$ represents not only the power-amplifier stage but also the control electronics section that provides drive signals.
2. At this time $Q_{3}$ and $Q_{4}$ are in the nonconducting (off) state. Turning $Q_{3}$ and $Q_{4}$ on (and $Q_{1}$ and $Q_{2}$ off) causes current to flow in the direction opposite to the arrow. Switching takes place continuously; at no time is the " H " bridge totally off-even when the motor is at rest.

With the motor at rest, current is delivered to it first in one direction, then in the other. The average current is then a triangular waveform centered on the zero axis as shown in Fig. 3.

While circulating current through the motor with no resulting motion seems wasteful, it prevents the servo system from exhibiting a dead zone at zero speed or "zero" position. As a result, both small and large mechanical disturbances can be corrected by the servo.

The cycle of events that generates the waveform of Fig. 3 further explains how the " H " bridge operates. When $Q_{1}$ and $Q_{2}$ are turned on by a driver in the control electronics, the supply voltage, $\mathrm{V}_{\mathrm{s}}$, appears across the inductor, L (Fig. 2). Current flows in the direction of the current arrow, $\mathrm{i}(\mathrm{t})$. (This is an ideal case, which assumes no saturation drops across the transistors, and zero motor back-emf.) Since $V_{s}=L(d i / d t)$, $i$ is a ramp of current as shown in Fig. 3 for the interval $0<t<t_{1}$.

Next $Q_{1}$ and $Q_{2}$ turn off while $Q_{3}$ and $Q_{4}$ turn on.

Current continues to flow in the same direction through L because of the inductor's collapsing field. Free-wheeling diodes $D_{3}$ and $D_{4}$ provide a continuous path for current flow. At the same time $a-V_{s}$ appears across L. This condition is shown in Fig. 3 for the interval $\mathrm{t}_{1} \leq \mathrm{t}<\mathrm{t}_{2}$.

At $t=t_{2}$, the inductor field totally collapses, and $Q_{3}$ and $Q_{\Delta}$ maintain $-V_{s}$ across $L$. A negative ramp of current occurs in the $\mathrm{t}_{2} \leq \mathrm{t} \leq \mathrm{t}_{3}$ period since di/dt $=-V_{s} / L$. Finally, $Q_{1}$ and $Q_{2}$ are again turned on, as $Q_{3}$ and $Q_{4}$ go off.
motor. A signal proportional to motor current is fed back to the control-electronics input. Operating the " H " bridge as a current amplifier rather than a voltage amplifier becomes important when an SSA is used in a position servo.

## Current sources are better

Sensing current in the " H " bridge and comparing it with the control electronics' input voltage produces a voltage-to-current transform. Current feedback,

2. Motor drive in an SSA is provided by an "H" bridge. Alternate pairs of switching transistors, $\mathrm{Q}_{1}$ through $\mathrm{Q}_{4}$. allow current to flow in the motor in two directions. An " $H$ " bridge is a current source that senses current flow

Even after $Q_{3}$ and $Q_{4}$ are off, a current of -I continues to flow through L on a path provided by $D_{1}$ and $D_{2}$-until the inductor field again collapses. The entire cycle then repeats itself.
Time delays are built into the amplifier to allow one transistor pair to turn off before the other pair comes on. This avoids a low impedance path from $\mathrm{V}_{\mathrm{s}}$ to ground through the transistors. In Fig. 2, Q1 must be fully off before $Q_{s}$ turns on, or $V_{s}$ will have a direct path to ground through them. The delays don't affect the waveforms because the diode pairs conduct immediately after a transistor pair is turned off. And the diode pairs allow enough time for recovery and turn on of the appropriate set of transistors.
In the emitter legs of $Q_{2}$ and $Q_{4}$ of Fig. 2, notice that two resistors sense current flow through the
through a motor and feeds it back to a summing junction at the control-electronics input. Current-source drive makes position servos stable and is a "universal" feature of all SSAs on the market.
from the current op amp in Fig. 2, is returned to the junction at the control-electronics input. There, it's summed with voltage from the output of the compensation network.

When a motor is driven from a current source, your servo system can benefit several ways:

- Position servos are stabilized by current drive, which results in a second-order system. On the other hand, voltage drive causes a third-order, or unstable system. (It doesn't matter what type of drive you use in a velocity servo, since it is naturally stable with either current or voltage sources. But since SSAs are designed as "universal" controllers, current drive is employed to ensure that every application is stable.)
- Current-limiting to the motor is easily provided. Clamping the input voltage of the power stage to a


3. Instantaneous current in a servo motor is a triangular waveform whose average value is zero. Pairs of switching transistors in an SSA's power amplifier drive current to the motor through a fixed inductor to generate a triangular waveform.

4. Servo-system performance can be modified by a compensation network. The open-loop transfer function (a) indicates an unstable system. A lag-lead network shifts $f_{c}$, the zero-dB crossover frequency, to a lower value, and restores stability to the system as shown in the modified transfer function (b).
maximum value sets the maximum current level, no matter what the input command dictates. And when the current limit is reached, the output transistors don't have to come out of saturation, which may occur with other types of current limiters.

- Inductor L , which is required in current-drive systems, smoothes the output of the " H " bridge and provides almost pure de to the motor (Fig. 2). A relatively large inductor (about 1 mH ) prevents current spikes-and protects the output transistors. Since current in L cannot change instantaneously, the inductor buys time, in effect, to detect faults and shut the system down before currents become excessive. Servo performance is not affected by an inductor as long as the output is operated as a current source.
- Short-circuits are protected against automatically. Current is delivered independently of load.

So great are the advantages of a current-source
output that virtually all SSAs are designed in this way. And reliability and ease of use make it a universal output stage.
But before a current-source power stage can drive a motor, your SSA must condition and control the input signals. To see how, examine the functions of the compensation network and control electronics.

## System control-points to ponder

In a high-gain servo, the zero- dB axis crossing of the open-loop transfer function can occur too close to the switching frequency, $f_{3}$; the result is an unstable system. Refer to Fig. 4a, where $f_{c}$ is the zero-dB crossover frequency. Here, $f_{c}$ and $f_{s}$ are close together; the result is a system that does not "seem continuous." This means that the electromechanical system, due to a large $f_{c}$, responds so quickly that significant mechanical motion occurs during a cycle of $\mathrm{f}_{\mathrm{s}}$. But if the amplifier switches much faster than the electromechanical system can respond, no significant motion will occur during $\mathrm{f}_{\mathrm{s}}$, and the system will "seem continuous."

To provide for a "continuous" servo, an SSA compensates with a lag-lead network. A modified transfer function due to compensation appears as shown in Fig. 4b. Compensating the amplifier causes $f_{c}$ to be less than or equal to $1 / 5 \mathrm{f}_{\mathrm{s}}$, so bandwidths close to $\mathrm{f}_{\mathrm{s}}$ are avoided and the servo is made continuous.

If system bandwidth isn't at most $1 / 5$ the minimum switching frequency, the system will act as a sampleddata servo and introduce poles that cause instability. But switching frequency is determined by the design of the control electronics, which can be one of two types. A constant-frequency controller always drives the " H " bridge at a frequency determined by a separate oscillator. A constant-ripple controller is a self-oscillator using a hysteresis switch. Since fewer problems are encountered with constant-frequency types, they're more widely used in SSAs. But at low system speeds, they can generate large ripple currents.

Because the back-emf of a motor operating at low speed is small, the voltage across the inductor is almost equal to the supply voltage. Therefore, $\mathrm{di} / \mathrm{dt}$ $=\mathrm{V}_{\mathrm{s}} / \mathrm{L}$ produces a large ripple current that depends on system speed as shown in Fig. 5. To see how large the ripple current can be, assume that an SSA has a switching frequency of 5 kHz . The period of the switching waveform is then $200 \mu \mathrm{~s}$, and in t/4 or 50 $\mu \mathrm{s}$, the waveform rises from zero to its maximum current. If $\mathrm{V}_{\mathrm{s}}$ is 75 V and the inductor is 1 mH , the rate of change of current will be
$\frac{\mathrm{di}}{\mathrm{dt}}=\frac{\mathrm{V}_{\mathrm{S}}}{\mathrm{L}}=\frac{75}{10^{-3}}=75 \times 10^{3} \mathrm{~A} / \mathrm{s}$,
and

$$
\Delta \mathrm{I}=\frac{\mathrm{di}}{\mathrm{dt}}(\Delta \mathrm{t})=75 \times 10^{3}\left(50 \times 10^{-6}\right)=3.75 \mathrm{~A}
$$

Peak currents of 3.75 A heat the motor and serve

5. Ripple current in a servo motor is a function of motor speed. With constant frequency control, ripple current is large at low motor speeds, increasing motor heating. But current must circulate through a motor at all times to prevent a system dead zone.

6. Ripple is constant, regardless of motor speed with a constant ripple controller. However, frequency variations can lead to instability in position servos.
no purpose other than to prevent a dead zone. But this problem can be overcome by constant-ripple control, which is designed to sense ripple and keep it independent of speed. Although ripple is fixed (Fig. 6), frequency varies depending on the speed of the motor, and this introduces other problems. When frequency varies, it can pass through a mechanical resonant point of the servo causing uncontrollable disturbances or loss of control. And because the switching frequency determines the upper limit of servo bandwidth, system bandwidth can vary. But as already shown, system bandwidth must be at least five times less than the minimum switching frequency for a system to be "continuous." Large frequency swings, therefore, can produce conditions that cause instability..e


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## Technology

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5. Determine the minimum effective volume.
6. Select the minimum-sized core.
7. Determine the number of turns.
8. Determine the air gap.
9. Check the incremental effect.
10. Determine the total power loss.
11. Determine the maximum operating temperature the inductor must withstand.

Follow this step-by-step method to determine the optimum magnetic characteristics without the usual lengthy iteration. The power dissipated in the coil's windings then becomes the major factor that limits the inductor's minimum size.

## Finding the coil current

But before you can start, distinguish between the maximum incremental current, $\mathrm{I}_{\mathrm{m}}$, and the rated current at maximum temperature, $I_{r}$. The incremental current is the total instantaneous ac or pulse current ( $\mathrm{I}_{\mathrm{pk}}$ ) and dc offset or bias ( $\mathrm{I}_{\mathrm{dc}}$ ):

$$
\begin{equation*}
\mathrm{I}_{\mathrm{m}}=\mathrm{I}_{\mathrm{pk}}+\mathrm{I}_{\mathrm{dc}} \tag{1}
\end{equation*}
$$

Since these currents together change the choke's inductance, the minimum inductance is determined by $\mathrm{I}_{\mathrm{m}}$.
The rated current is the total effective ac, $\mathrm{L}_{\text {eff, }}$ and $\mathrm{I}_{\mathrm{dc}}$ :

$$
\begin{equation*}
I_{r}=I_{e f f}+I_{d c} \tag{2}
\end{equation*}
$$

These currents heat the inductor, so $I_{r}$ determines the minimum size of the magnet wire.

[^2]

1. A trapezoid of current (a) results when a square wave of voltage (b) is impressed on an inductor. The initial jump of current, $I_{c}$, ramps up to the final value, $I_{p k}$, with a slope that depends on coil inductance and resistance.

With a sine wave of $\mathrm{E}_{\mathrm{ac}}$, the maximum rms voltage across the choke, $\mathrm{I}_{\text {eff }}$ depends on $\mathrm{L}_{\mathrm{m}}$, the incremental inductance at $I_{m}$, and the operating frequency, $f$ :

$$
\begin{equation*}
\mathrm{L}_{\mathrm{ff}}=\mathrm{E}_{\mathrm{ac}} / 2 \pi \mathrm{fL}_{\mathrm{m}} \tag{3}
\end{equation*}
$$

also

$$
\begin{equation*}
I_{p k}=\sqrt{2} I_{\text {eff }} \tag{4}
\end{equation*}
$$

Be careful in assigning a value to f -it may not be the line frequency. For instance, with a $60-\mathrm{Hz}$ threephase full-wave-bridge rectifier, $f$ is 360 Hz .
For pulse excitation of the choke, as in switching regulators, the voltage and current waveforms are shown in Fig. 1. When a choke handles pulses rather than sine waves, $\mathrm{I}_{\mathrm{pk}}$ depends on the pulse excitation voltage ( $\mathrm{E}_{\mathrm{pk}}$ ), the pulse duration ( $\mathrm{D}_{\mathrm{p}}$ ), the $\mathrm{L}_{\mathrm{m}}$ and the core-loss current ( $\mathrm{I}_{\mathrm{c}}$ ), which is due to hysteresis and eddies:

$$
\begin{equation*}
\mathrm{I}_{\mathrm{pk}}=\left(\mathrm{E}_{\mathrm{pk}} \mathrm{D}_{\mathrm{p}} / \mathrm{L}_{\mathrm{m}}\right)+\mathrm{I}_{\mathrm{c}} . \tag{5}
\end{equation*}
$$

Choose the proper core and you can keep $I_{c}$ less than $10 \%$ of $\mathrm{I}_{\mathrm{pk}}$ :

$$
\begin{equation*}
\mathrm{I}_{\mathrm{c}} \leq 0.1 \mathrm{I}_{\mathrm{pk}} \tag{5a}
\end{equation*}
$$

The pulse-repetition time, $\mathrm{t}_{\mathrm{pr}}$, comes into play for $\mathrm{I}_{\text {eff. }}$.

$$
\begin{equation*}
\mathrm{L}_{\mathrm{eff}}=\sqrt{\mathrm{D}_{\mathrm{p}}\left(\mathrm{I}_{\mathrm{p}}^{2}+\mathrm{I}_{\mathrm{pk}} \mathrm{I}_{\mathrm{c}}+\mathrm{I}_{\mathrm{c}}^{2}\right) / 3 \mathrm{t}_{\mathrm{pr}}} \tag{6}
\end{equation*}
$$

Once you've determined the values for the ap-

## Table 1. Average values of common core materials.

| Material <br> Designation | Freq. Range (Hz) | Temp. Range $\left({ }^{\circ} \mathrm{C}\right)$ | $\begin{gathered} U_{i} \\ (G / O e) \end{gathered}$ | Bs <br> (G) | $B_{r}$ <br> (G) | $\begin{gathered} \mathbf{H}_{\mathbf{i}} \\ (\mathrm{Oe}) \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Electrical Steels ${ }^{4}$ |  |  |  |  |  |  |
| Silicon Iron | 20 to 1 k | -55 to +300 | 500 | 17.5 k | 12 k | 5.5 |
| Silectron | 20 to 10 k | -55 to +375 | 1.5 k | 16.5 k | 14 k | 0.83 |
| Alloy 48 | 20 io 8 k | -55 to +250 | 1 k | 12.5 k | 10 k | 1.25 |
| HY-MV 80 | 20 to 25 k | -55 to +230 | 10 k | 8 k | 4.4 k | 0.18 |
| Supermalloy | 20 to 25 k | -55 to +230 | 50 k | 7.3 k | 4 k | 0.033 |
| Supermendur | 20 to 2 k | -55 to +460 | 1.5 k | 21 k | 19 k | 0.67 |
| Molypermalloy Powders |  |  |  |  |  |  |
| MPP 14² | 400 kto 1 M | -55 to +250 | 14 | 6 k | 10 | 214 |
| MPP $26{ }^{2}$ | 400 to 650 k | -55 to +230 | 26 | 6 k | 10 | 115 |
| MPP 60 | 400 to 250 k | -55 to +200 | 60 | 6 k | 10 | 49.8 |
| MPP 125 | 40 to 100 k | -55 to +200 | 125 | 6 k | 200 | 23.2 |
| MPP 160 | 40 to 70 k | -55 to +200 | 160 | 6 k | 200 | 18.1 |
| MPP 200 | 40 to 30 k | -55 to +175 | 200 | 6 k | 200 | 14.5 |
| MPP 300 | 40 to 25 k | -55 to +150 | 300 | 6 k | 300 | 9.5 |
| MPP 550 | 40 to 20 k | -55 to +125 | 550 | 6 k | 650 | 4.86 |
| Powdered Irons |  |  |  |  |  |  |
| Carbonyl SF2 | 2 to 50 M | -55 to +125 | 7.5 | 8 k | 10 | 533 |
| Carbonyl E2 | 200 k to 10 M | -55 to +125 | 10 | 8 k | 10 | 400 |
| Carbonyl C ${ }^{2}$ | 100 k to 2 M | -55 to +125 | 20 | 8 k | 10 | 200 |
| Carbonyl GQ4² | 50 k to 1 M | -55 to +125 | 35 | 8 k | 10 | 114 |
| Carbonyl HA | 1 to 100 k | -55 to +105 | 60 | 8 k | 2200 | 48.3 |
| 75 Powder | 400 to 50 k | -55 to +105 | 75 | 8 k | 2200 | 38.7 |
| 90 Powder | 400 to 10 k | -55 to +105 | 90 | 8 k | 2200 | 32.2 |
| Ferrites ${ }^{3}$ |  |  |  |  |  |  |
| F40 | 10 to 80 M | -55 to +250 | 40 | 2400 | 750 | 20.6 |
| F125 | 200 k to 10 M | -55 to +250 | 125 | 2350 | 1200 | 4.6 |
| F175 | 100 k to 5 M | -55 to +230 | 175 | 2550 | 1400 | 3.28 |
| F250 | 50 k to 4 M | -55 to +200 | 250 | 2200 | 1100 | 2.2 |
| F400 | 10 k to 2.5 M | -55 to +175 | 400 | 2700 | 1100 | 2 |
| F750 | 1 k to 1.5 M | -55 to +125 | 750 | 4000 | 1800 | 1.47 |
| F1000 | 1 k to 1 M | -55 to +125 | 1000 | 4200 | 1700 | 1.25 |
| F1500 | 1 k to 650 k | -55 to +125 | 1500 | 4000 | 1100 | 0.97 |
| F2000 | 400 to 500 k | -55 to +125 | 2000 | 4400 | 1500 | 0.72 |
| F2300 | 400 to 300 k | -40 to +105 | 2300 | 4000 | 1200 | 0.61 |
| F2700 | 400 to 250 k | -30 to +105 | 2700 | 4700 | 2000 | 0.50 |
| F5000 | 400 to 100 k | -25 to +105 | 5000 | 4300 | 1200 | 0.31 |
| F10000 | 400 to 80 k | -25 to +90 | 10000 | 4300 | 1200 | 0.16 |

## Notes:

(1) For specific data and tolerances, refer to individual suppliers. (2) Excessive heating occurs before incremental effects are observed. (3) There are no standard designations for equivalent materials between ferrite suppliers. Designation used is for convenience. (4) Frequency range of electrical steel depends on material thickness. The thinner the material. the higher the frequency range

| Symbol | Terminology | Units |
| :---: | :---: | :---: |
| A | Cross-sectional area of the core | $\mathrm{cm}^{2}$ |
| A Lo | Initial-inductance index | H/turn ${ }^{2}$ |
| A Lm | Incremental-inductance index | H/turn ${ }^{2}$ |
| $A_{\text {s }}$ | Total surface area | $\mathrm{cm}^{2}$ |
| B | Maximum operating-flux density | G |
| $\mathrm{Br}_{\text {r }}$ | Residual-flux density | G |
| $\mathrm{B}_{5}$ | Saturation-flux density | G |
| $\mathrm{C}_{\text {v }}$ | Core loss per unit volume | W/cm ${ }^{3}$ |
| $C_{\text {vf }}$ | Core-loss factor | $\mathrm{W} / \mathrm{cm}^{3} / \mathrm{Hz}$ |
| $\mathrm{d}_{\mathrm{i}}$ | Insulated-wire diameter | in. |
| ${ }^{\text {d }}$ o | Bare-wire diameter | in. |
| D p | Pulse duration | s |
| E ac | Ac-excitation voltage | rms V |
| Epk | Pulse-excitation voltage | pk V |
| , | Design operating frequency | Hz |
| $\mathrm{G}_{\mathrm{t}}$ | Thermal conductance | W/ ${ }^{\circ} \mathrm{C}$ |
| H | Maximum operating magnetic-field strength | Oe |
| $\mathrm{H}_{\mathrm{i}}$ | Intrinsic magnetic-field strength | Oe |
| $\mathrm{I}_{\mathrm{c}}$ | Core-loss current | A |
| I dc | Dc current (offset or bias) | A |
| $I_{\text {eff }}$ | Effective current | A |
| $\mathrm{If}_{\mathrm{f}}$ | Fusing current | A |
| 1 m | Maximum incremental current | A |
| I pk | Pulse current | A |
| $\mathrm{I}_{\mathrm{r}}$ | Rated current at max temperature | A |
| I ro | Rated current at room temperature | A |
| k | Winding-utilization factor | --- |
| L | Inductance | H |
| L c | Magnetic-path length of the core | cm |
| 1 e | Effective magnetic-path length | cm |
| 1 g | Total air-gap length | cm |
| L m | Incremental inductance at I m | H |
| 1 tm | Mean length per turn | ft |
| Lo | Initial inductance | H |
| N | Total turns | turns |
| p | Dc resistance per unit length | $\Omega / \mathrm{ft}$ |
| $\mathrm{P}_{\mathrm{c}}$ | Core loss | W |
| $\mathrm{P}_{\text {t }}$ | Total power | w |
| $\mathrm{P}_{\mathrm{w}}$ | Winding loss | W |
| r | Thermal coefficient of resistance | $\Omega / \Omega /{ }^{\circ} \mathrm{C}$ |
| R dc | Dc winding resistance | $\Omega$ |
| $\mathrm{S}_{\mathrm{d}}$ | Surface dissipation | W/cm ${ }^{2} /{ }^{\circ} \mathrm{O}$ |
| T a | Ambient temperature | ${ }^{\circ} \mathrm{C}$ |
| T m | Maximum operating temperature | ${ }^{\circ} \mathrm{C}$ |
| To | Reference temperature | ${ }^{\circ} \mathrm{C}$ |
| T pr | Pulse-repetition time | $s$ |
| Tr | Temperature rise | ${ }^{\circ} \mathrm{C}$ |
| ${ }^{\text {t }} \mathrm{p}^{2}{ }^{\text {a }}$ | Winding density | turns/in. ${ }^{2}$ |
| $U_{\text {e }}$ | Effective permeability | G/Oe |
| $u_{i}$ | Initial permeability | G/Oe |
| $\mathrm{U}_{\mathrm{m}}$ | Incremental permeability | G/Oe |
| $\mathrm{V}_{\mathrm{e}}$ | Effective volume | $\mathrm{cm}^{3}$ |
| V em | Minimum effective volume | $\mathrm{cm}^{3}$ |
| $V_{c}$ | Actual core volume | $\mathrm{cm}^{3}$ |
| $\mathrm{W}_{\text {a }}$ | Winding cross-sectional area | $\mathrm{cm}^{3}$ |

propriate coil currents, you can select a core material from the hundreds of magnetic materials available. Table 1 lists a representative cross-section of several popular core materials.

To pick the right core material, you need to know the required $f$, maximum operating temperature, $\mathrm{T}_{\mathrm{m}}$, core properties you need, cost, geometry and winding limitations, among other things. But whenever possible, and especially for switching regulators, use gapped-ferrite cores.

The core loss per unit volume, $\mathrm{C}_{\mathrm{v}}$, depends on the core-loss factor:

$$
\begin{equation*}
\mathrm{C}_{\mathrm{v}}=\mathrm{C}_{\mathrm{vf}} \mathrm{f} \tag{7}
\end{equation*}
$$

The crux of this entire procedure is that it determines explicitly-without iteration-optimum values for maximum operating flux density, B, and maximum operating magnetic-field strength, H , for the selected core material. Empirical analysis has shown that incremental effects begin approximately when the intrinsic magnetic-field strength, $\mathrm{H}_{\mathrm{i}}$, equals the saturation-flux density, $\mathrm{B}_{\mathrm{s}}$, minus the residualflux density, $\mathrm{B}_{\mathrm{r}}$, all divided by twice the initial permeability, $\mathrm{U}_{\mathrm{i}}$, or

$$
\begin{equation*}
\mathrm{H}_{\mathrm{i}}=\left(\mathrm{B}_{\mathrm{s}}-\mathrm{B}_{\mathrm{r}}\right) / 2 \mathrm{U}_{\mathrm{i}} \tag{8}
\end{equation*}
$$

Thus, for inductors passing substantial de, you must lower the maximum operating flux density to the differential flux density, $B_{s}-B_{r}$, rather than just $B_{s}$ :

$$
\begin{equation*}
\mathrm{B}=\mathrm{B}_{\mathrm{s}}-\mathrm{B}_{\mathrm{r}} . \tag{9}
\end{equation*}
$$

Maximum operating field strength depends not only on $B$, but also on incremental permeability, $\mathrm{U}_{\mathrm{m}}$ :

$$
\begin{equation*}
\mathrm{U}_{\mathrm{m}}=\mathrm{U}_{\mathrm{i}} \mathrm{~L}_{\mathrm{m}} / \mathrm{L}_{0}, \tag{10}
\end{equation*}
$$

where $L_{0}$ is the choke's initial low-level ac value of inductance with no dc, and $L_{m}$ is the desired inductance value at $\mathrm{I}_{\mathrm{m}}$.

Now you can find H :

$$
\begin{equation*}
\mathrm{H}=\mathrm{B} / \mathrm{U}_{\mathrm{m}} . \tag{11}
\end{equation*}
$$

Because of the large variations in the permeabilities of many magnetic materials, especially the electrical steels, specify $\mathrm{U}_{\mathrm{i}}$ at no more than 40 G . And expect variations in $\mathrm{U}_{\mathrm{i}}, \mathrm{B}_{\mathrm{s}}$ and $\mathrm{B}_{\mathrm{r}}$ at high temperatures.

## Minimizing magnet-wire size

Having found values for B and H , you can figure out the thinnest magnet wire that you can use.

The size of the magnet wire is limited by its current rating, $I_{r}$ However, the usual current rating for copper wire doesn't apply to inductors. Ratings based on a current density of 1000 circular mils/A-the loading that causes a $2 \%$ voltage drop per 100 ft of standard house wiring-aren't realistic.
Instead current ratings for industrial and military inductors are based on a more practical consideration -maximum temperature rise. Since these ratings range typically from 5 to $20 \%$ of the wire's fusing current, rating the inductor wire at $10 \%$ of its fusing current is realistic, not to mention convenient for calculation. This rating applies at $\mathrm{T}_{\mathrm{m}}$.

At 20 C , the fusing current for bare-copper magnet wire of diameter $d_{0}$ is found by

Table 2. Magnet Wire Design Data.

|  | Bare Wire |  |  |  | Single Film |  |  |  |  | Double Film |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { AWG } \\ & \text { Size } \end{aligned}$ | Max 00 | Max I | Max Tensile Strength | $\Omega / \mathrm{tt}$ | Max 00 | f1/lb | ת/1b | Tpi | Tpi ${ }^{2}$ | Max 00 | ft/lb | ת/1b | $\mathrm{T}_{\mathrm{pi}}$ | $T_{\mathrm{pi}}{ }^{2}$ |
| 10 | 0.1024 | 336 | 82.4 | 000100 | 0.1047 | 316 | 0.0316 | 9551 | 91.22 | 0.1061 | 315 | 00315 | 9425 | 88.83 |
| 11 | 0.0912 | 28.2 | 653 | 000126 | 0.0935 | 398 | 0.0501 | 10.70 | 1144 | 0.0948 | 39.7 | 00500 | 1055 | 111.3 |
| 12 | 00812 | 237 | 51.8 | 0.00159 | 0.0834 | 503 | 0.0800 | 11.99 | 1438 | 00847 | 500 | 0.0795 | 11.81 | 139.4 |
| 13 | 0.0724 | 199 | 412 | 0.00200 | 0.0746 | 633 | 0.1266 | 13.40 | 1797 | 0.0757 | 629 | 0.1258 | 13.21 | 1745 |
| 14 | 0.0644 | 16.7 | 326 | 000252 | 0.0666 | 79.9 | 0.2013 | 15.02 | 2255 | 00682 | 793 | 0.1998 | 14.66 | 2150 |
| 15 | 00574 | 141 | 259 | 000318 | 0.0594 | 101 | 03212 | 1684 | 283.4 | 00609 | 100 | 0.3180 | 16.42 | 2696 |
| 16 | 0.0511 | 11.8 | 20.5 | 0.00402 | 0.0531 | 127 | 05105 | 1883 | 354.7 | 00545 | 126 | 05065 | 18.35 | 336.7 |
| 17 | 00455 | 994 | 163 | 000505 | 00475 | 159 | 08029 | 2105 | 4432 | 0.0488 | 158 | 0.7979 | 2049 | 4199 |
| 18 | 00405 | 8.35 | 129 | 0.00639 | 0.0424 | 201 | 1284 | 23.58 | 556.2 | 00437 | 199 | 1272 | 22.88 | 523.6 |
| 19 | 0.0361 | 7.02 | 10.2 | 0.00805 | 0.0379 | 253 | 2037 | 2639 | 696.2 | 00391 | 251 | 2.020 | 2558 | 654.1 |
| 20 | 0.0322 | 5.92 | 8.14 | 00101 | 00339 | 318 | 3.212 | 2950 | 870.2 | 0.0351 | 315 | 3.182 | 28.49 | 8117 |
| 21 | 0.0286 | 495 | 642 | 0.0128 | 0.0303 | 402 | 5146 | 3300 | 1089 | 0.0314 | 397 | 5082 | 31.85 | 1014 |
| 22 | 0.0254 | 415 | 507 | 00162 | 00270 | 508 | 8.230 | 3704 | 1372 | 0.0281 | 503 | 8.149 | 3559 | 1266 |
| 23 | 0.0227 | 350 | 405 | 00203 | 00243 | 633 | 12.85 | 41.15 | 1694 | 00253 | 625 | 1269 | 39.53 | 1562 |
| 24 | 00202 | 294 | 320 | 0.0257 | 00217 | 806 | 2071 | 46.08 | 2124 | 0.0227 | 794 | 20.41 | 4405 | 1941 |
| 25 | 0.0180 | 2.47 | 254 | 0.0324 | 0.0194 | 1013 | 32.82 | 51.55 | 2657 | 00203 | 990 | 3208 | 4926 | 2427 |
| 26 | 0.0160 | 207 | 2.01 | 0.0410 | 00173 | 1282 | 5256 | 5780 | 3341 | 00182 | 1260 | 5166 | 5495 | 3019 |
| 27 | 00143 | 175 | 161 | 0.0514 | 0.0156 | 1608 | 8265 | 64.10 | 4109 | 00164 | 1580 | 8121 | 60.98 | 3718 |
| 28 | 00127 | 147 | 1.27 | 0.0653 | 00140 | 2033 | 1328 | 71.43 | 5102 | 00147 | 1990 | 129.9 | 68.03 | 4628 |
| 29 | 0.0114 | 125 | 1.02 | 00812 | 0.0126 | 2525 | 2050 | 79.37 | 6299 | 00133 | 2470 | 2006 | 75.19 | 5653 |
| 30 | 0.0101 | 1.04 | 0801 | 0104 | 00112 | 3215 | 3344 | 89.29 | 7972 | 00119 | 3140 | 3266 | 84.03 | 7062 |
| 31 | 00090 | 0.874 | 0636 | 0.131 | 00100 | 4065 | 5325 | 1000 | 10000 | 00108 | 3950 | 5174 | 9259 | 8573 |
| 32 | 00081 | 0.747 | 0.515 | 0.162 | 00091 | 5000 | 810.0 | 1099 | 12076 | 00098 | 4880 | 7906 | 102.0 | 10412 |
| 33 | 00072 | 0.626 | 0407 | 0206 | 0.0081 | 6369 | 1312 | 1235 | 15242 | 00088 | 6170 | 1271 | 113.6 | 12913 |
| 34 | 00064 | 0.524 | 0.322 | 0.261 | 0.0072 | 8064 | 2105 | 138.9 | 19290 | 0.0078 | 7870 | 2054 | 128.2 | 16437 |
| 35 | 0.0057 | 0441 | 0.255 | 0.331 | 00064 | 10210 | 3380 | 156.2 | 24414 | 00070 | 9940 | 3290 | 142.9 | 20408 |
| 36 | 0.0051 | 0.373 | 0.204 | 0.415 | 00058 | 12760 | 5295 | 172.4 | 29727 | 00063 | 12440 | 5163 | 158.7 | 25195 |
| 37 | 0.0046 | 0.319 | 0.166 | 0.512 | 0.0052 | 15800 | 8090 | 192.3 | 36982 | 00057 | 15300 | 7834 | 1754 | 30779 |
| 38 | 0.0041 | 0.269 | 0.132 | 0.648 | 00047 | 19920 | 12908 | 2128 | 45269 | 0.0051 | 19300 | 12506 | 196.1 | 38447 |
| 39 | 00036 | 0221 | 0101 | 0.847 | 0.0041 | 26040 | 22056 | 2439 | 59488 | 0.0045 | 25100 | 21260 | 222.2 | 49383 |
| 40 | 00032 | 0.185 | 00804 | 1.08 | 00037 | 33110 | 35759 | 2703 | 73046 | 0.0040 | 32200 | 34776 | 250.0 | 62500 |
| 41 | 0.0029 | 0.160 | 00661 | 132 | 0.0033 | 40100 | 52932 | 3030 | 91827 | 00036 | 39500 | 52140 | 2778 | 77160 |
| 42 | 0.0026 | 0136 | 00531 | 166 | 00030 | 51000 | 84660 | 3333 | 111111 | 0.0032 | 49800 | 82668 | 3125 | 97656 |
| 43 | 0023 | 0113 | 00475 | 2.14 | 0.0026 | 65800 | 140.8k | 384.6 | 147928 | 00029 | 63700 | 136.3k | 3448 | 118906 |
| 44 | 0.0021 | 00985 | 00346 | 259 | 00024 | 79400 | 205.6 k | 4167 | 173611 | 00027 | 76300 | 1976k | 3704 | 137174 |
| 45 | 000176 | 00756 | 00243 | 362 | 000205 | 104k | 376.5k | 4878 | 237954 | 000230 | 99600 | 3606k | 434.8 | 189036 |
| 46 | 000157 | 0.0637 | 00194 | 4.54 | 0.00185 | 132k | 599.3k | 5405 | 292184 | 000210 | 126k | 572 0k | 4762 | 226757 |
| 47 | 000140 | 00536 | 0.0154 | 571 | 000170 | 162k | 9250k | 5882 | 346020 | 000190 | 153k | 8736k | 5263 | 277003 |
| 48 | 0.00124 | 00447 | 0.0121 | 729 | 000150 | 205k | 1494 M | 6666 | 444444 | 000170 | 199k | 1451 M | 5882 | 346020 |
| 49 | 000111 | 00379 | 00097 | 909 | 000130 | 258k | 2345 M | 7692 | 591716 | 0.00150 | 252k | 2.291 M | 6666 | 444444 |
| 50 | 000099 | 0.0319 | 00077 | 11.4 | 000120 | 312k | 3557M | 8333 | 694444 | 0.00140 | 306k | 3 488M | 7143 | 510204 |
| 51 | 0.00088 | 00267 | 00061 | 145 | 000110 | 416k | 6032M | 909.1 | 826446 |  |  |  |  |  |
| 52 | 0.00078 | 00223 | 00048 | 184 | 000100 | 555k | 1021M | 1000 | 1000 M |  |  |  |  |  |
| 53 | 000070 | 00190 | 00038 | 229 | 0.00085 | 667k | 1527M | 1176 | 1384 M |  |  |  |  |  |
| 54 | 000062 | 00158 | 00030 | 29.1 | 0.00075 | 859k | 2500 M | 1333 | 1777M |  |  |  |  |  |
| 55 | 000055 | 00132 | 00024 | 37.0 | 0.00070 | 1090M | 4033 M | 1429 | 2041M |  |  |  |  |  |
| 56 | 000049 | 00111 | 00019 | 466 | 000065 | 1380M | 6431 M | 1538 | 2367 M |  |  |  |  |  |
| Units | Inches | Amperes | Pounds | $\Omega / 4$ | Inches | 17/16 | $\Omega / 10$ | $T_{p i}$ | $\mathrm{T}_{\mathrm{pi}}{ }^{2}$ | Inches | tt/1b | ת/1b | Tpi | Tpir |

Maximum ODs $\Omega / / 11$ and $\mathrm{ft} / \mathrm{lb}$ are taken Irom Materials and Processes Handbook Maximum rated current is $10^{\circ}$ of the lusing current at 20 C Maximum tension is based upon a tensile strength of $1000 \mathrm{PSI} \Omega / \mathrm{lb}$ is delived from $(\Omega / \mathrm{fl} \times(1 / / \mathrm{lb}) \mathrm{Tpl} \quad 1 / \mathrm{Max} 00 \mathrm{TpI} \quad 1 / \mathrm{Max} 0 \mathrm{OD}$

$$
\begin{equation*}
\mathrm{I}_{\mathrm{f}}=10,244 \mathrm{~d}_{0}{ }^{3} \tag{12}
\end{equation*}
$$

As a result, the current rating at 20 C is

$$
\begin{equation*}
\mathrm{I}_{\mathrm{ro}}=0.1 \mathrm{I}_{\mathrm{f}} \text {, } \tag{13}
\end{equation*}
$$

Furthermore, for a required $I_{r}$, the magnet wire's minimum current rating at 20 C must be

$$
\begin{equation*}
\mathrm{I}_{\mathrm{ro}}=\mathrm{I}_{\mathrm{r}}\left[1+0.00393\left(\mathrm{~T}_{\mathrm{m}}-20\right)\right] \tag{14}
\end{equation*}
$$

Current ratings for copper magnet wire, based on Eq. 13, are tabulated in Table 2 along with other selected magnet-wire design data.

## Finding the effective volume

Using your values for B and H , determine the minimum effective volume, $\mathrm{V}_{\mathrm{em}}$, that will sustain your inductor's operating conditions.

$$
\begin{equation*}
\mathrm{V}_{\mathrm{em}}=0.4 \pi \times 10^{8} \mathrm{~L}_{\mathrm{m}} \mathrm{I}_{\mathrm{m}}^{2} /(\mathrm{B} \mathrm{H}) \tag{15}
\end{equation*}
$$

For ungapped cores like toroids, the actual core volume must equal or exceed $\mathrm{V}_{\mathrm{em}}$. But meeting this requirement may make the core size excessive. Fortunately, however, an air gap in the magnetic circuit can reduce core size significantly. If the air-gap length, $\mathrm{l}_{\mathrm{g}}$, is small with respect to the effective magnetic-path length, $l_{e}$, the equivalent effective volume, $V_{e}$, is:

$$
\begin{align*}
\mathrm{V}_{\mathrm{e}} & =A_{\mathrm{c}} l_{\mathrm{e}} \\
& =A_{c}\left(l_{\mathrm{c}}+U_{i} l_{\mathrm{g}}\right) \\
& =\mathrm{V}_{\mathrm{c}}+\left(\mathrm{A}_{\mathrm{c}} \mathrm{U}_{\mathrm{i}} \mathrm{l}_{\mathrm{g}}\right), \tag{16}
\end{align*}
$$

where $l_{e}$ is the magnetic path length of the core, $\mathrm{A}_{\mathrm{c}}$ is the core's cross-sectional area and $\mathrm{V}_{\mathrm{c}}$ is the actual volume.

To get the $L_{m}$ when you select the core gap ensure

$$
\begin{equation*}
\mathrm{V}_{\mathrm{e}} \geq \mathrm{V}_{\mathrm{em}} \tag{17}
\end{equation*}
$$

Armed with the minimum effective core volume, select the smallest core that can sustain the required inductance. With the help of core-size indexes pick a core from magnetic-core catalogs. They also provide data for the winding area, $W_{a}$, and for $A_{c}$ that you can use in

$$
\begin{equation*}
\mathrm{W}_{\mathrm{a}} \mathrm{~A}_{\mathrm{c}}=5.067 \times 10^{8} \mathrm{~L}_{\mathrm{m}} \mathrm{I}_{\mathrm{m}} \mathrm{~d}_{\mathrm{i}}{ }^{2} /(\mathrm{k} \mathrm{~B}) \tag{18}
\end{equation*}
$$

Here $\mathrm{d}_{\mathrm{i}}$ is the diameter of the insulated-wire (doublefilm insulation is recommended), and k is the windingutilization, or "fit," factor. For toroidal windings, k is typically 0.4 ; for bobbin windings, 0.8 .

The core must satisfy Eqs. 15 and 18. Often, the core's geometry is severely limited by mounting space and dc winding resistance, $\mathrm{R}_{\mathrm{dc}}$. (Other factors that affect core geometry are compared in Table 3.)

Even slug-type inductors can be designed with this step-by-step procedure, but you must empirically determine $\mathrm{l}_{\mathrm{g}}$ and $\mathrm{V}_{\mathrm{e}}$ for slugs.

Now that you know the material, size and geometry of the core, you can calculate the maximum value of the incremental inductance index, $\mathrm{A}_{\mathrm{Lm}}$, from

$$
\begin{equation*}
A_{\mathrm{Lm}}=\left(\mathrm{B}_{\mathrm{c}}\right)^{2} \times 10^{-16} /\left(\mathrm{L}_{\mathrm{m}} \mathrm{I}_{\mathrm{m}}{ }^{2}\right) . \tag{19}
\end{equation*}
$$

Then with $\mathrm{A}_{\mathrm{Lm}}$, compute the number of turns, N , required for the $L_{m}$ you need:

$$
\begin{equation*}
\mathrm{N}=\sqrt{\mathrm{L}_{\mathrm{m}} / \mathrm{A}_{\mathrm{Lm}}} \tag{20}
\end{equation*}
$$

Take N and H , and determine $\mathrm{l}_{\mathrm{e}}$ :

$$
\begin{equation*}
\mathrm{l}_{\mathrm{e}}=0.4 \pi \mathrm{NI}_{\mathrm{m}} / \mathrm{H} . \tag{21}
\end{equation*}
$$

For cores without air gaps, such as toroids, make sure that

Table 3. Core-geometry selection criteria.

| Selection <br> Factor | E-U-I <br> Cores | Pot <br> Cores | Toroid <br> Cores | Slug <br> Cores | Other <br> Shielded <br> Cores |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Core Cost | Low | High | Low | Low | High |
| Winding <br> Cost | Low | Low | High | Low | Low |
| Winding <br> Flexibility | Excellent | Fair | Good | Good | Fair |
| Mounting <br> Flexibility | Good | Good | Fair | Fair | Good |
| Shielding | Fair | Excellent | Good | Poor | Good |

$$
\begin{equation*}
l_{c} \geq 1_{e} . \tag{22}
\end{equation*}
$$

For cores with air gaps, compute the air-gap length from

$$
\begin{equation*}
\mathrm{l}_{\mathrm{g}}=\left(\mathrm{l}_{\mathrm{e}}-\mathrm{l}_{\mathrm{c}}\right) / \mathrm{U}_{\mathrm{i}} \tag{23}
\end{equation*}
$$

Now you can find the effective permeability:

$$
\begin{equation*}
\mathrm{U}_{\mathrm{e}}=\mathrm{U}_{\mathrm{i}} /\left[1+\left(\mathrm{U}_{\mathrm{i}} \mathrm{l}_{\mathrm{g}} / \mathrm{l}_{\mathrm{c}}\right)\right] . \tag{24}
\end{equation*}
$$

For core geometries in which an air gap interrupts the magnetic circuit twice, the inserted material thickness should be half that computed in Eq. 23.
The air gap affects the initial-inductance index, $A_{L 0}$, as follows:

$$
\begin{gather*}
\mathrm{A}_{\mathrm{Lo}}=0.4 \pi \times 10^{-8} \mathrm{~A}_{\mathrm{c}} /\left[\left(\mathrm{l}_{\mathrm{c}} / \mathrm{U}_{\mathrm{i}}\right)+\mathrm{l}_{\mathrm{g}}\right] .  \tag{25}\\
\text { Calculate the initial inductance, } \mathrm{L}_{0} \text {, from } \\
\mathrm{L}_{0}=\mathrm{A}_{\mathrm{Lo}} \mathrm{~N}^{2} . \tag{26}
\end{gather*}
$$

Next, compute the percent change in inductance $L$

$$
\begin{equation*}
\% \mathrm{~L}=\left[\left(\mathrm{L}_{\mathrm{m}} / \mathrm{L}_{0}\right)-1\right] 100 \tag{27}
\end{equation*}
$$

Don't let the inductance change more than $25 \%$. Next, compute $\mathrm{P}_{\mathrm{t}}$, the total power dissipated by the complete inductor-both the winding and the core. First determine the power dissipated in the windings, $\mathrm{P}_{\mathrm{w}}$.
The $d_{i}$ used for the magnet wire when finding the minimum-sized core is the minimum for the $\mathrm{I}_{\mathrm{r}}$. However, the $\mathrm{W}_{\mathrm{a}}$ of the actual core may be able to accept a wire with a large diameter and thereby lower the windings' power loss. The maximum diameter for the insulated wire is related to the winding density in turns per inch, tpi ${ }^{2}$, by

$$
\begin{equation*}
\mathrm{d}_{\mathrm{i}}=\sqrt{1 / \mathrm{tpi}^{2}} \tag{28}
\end{equation*}
$$

The winding density must conform to

$$
\begin{equation*}
\operatorname{tpi}^{2}=6.452 \mathrm{~N} /\left(\mathrm{k} \mathrm{~W}_{\mathrm{a}}\right) . \tag{29}
\end{equation*}
$$

Select a wire with the best size for lowest dc resistance per unit length, $p$. The de winding resistance, for a particular mean length per turn, $\mathrm{l}_{\mathrm{tm}}$, is found with

$$
\begin{equation*}
\mathrm{R}_{\mathrm{dc}}=\mathrm{N}_{\mathrm{tm}} \mathrm{p} \tag{30}
\end{equation*}
$$

Compute $\mathrm{P}_{\mathrm{w}}$, using $\mathrm{R}_{\mathrm{dc}}$, the reference temperature $T_{0}$, the wire's thermal coefficient of resistance ( $r$ ), and the $T_{m}$ and $I_{r}$ :

$$
\begin{equation*}
P_{\mathrm{w}}=\mathrm{I}_{\mathrm{r}} \mathrm{R}_{\mathrm{dc}}\left[1+\mathrm{r}\left(\mathrm{~T}_{\mathrm{m}}-\mathrm{T}_{\mathrm{o}}\right)\right] \tag{31}
\end{equation*}
$$

If $\mathrm{T}_{\mathrm{o}}$ is $20 \mathrm{C}, \mathrm{r}$ is 0.00393 ; if $\mathrm{T}_{0}$ is $25 \mathrm{C}, \mathrm{r}$ is 0.00385 .
Now, determine the core loss, $\mathrm{P}_{\mathrm{c}}$, from the data for
core loss per unit volume, $C_{v}$, which most core manufacturers either graph or tabulate. If necessary, you can extrapolate the value of $\mathrm{C}_{\mathrm{v}}$ for your conditions:

$$
\begin{equation*}
\mathrm{P}_{\mathrm{c}}=\mathrm{C}_{\mathrm{v}} \mathrm{~V}_{\mathrm{c}} . \tag{32}
\end{equation*}
$$

Finally, the total power dissipated by the inductor

$$
\begin{equation*}
P_{t}=P_{w}+P_{c} \tag{33}
\end{equation*}
$$

Of course, since you haven't built your inductor yet you can only estimate its maximum operating temperature. Surface dissipation, $\mathrm{S}_{\mathrm{d}}$, is the wide-ranging variable that makes determining the $\mathrm{T}_{\mathrm{m}}$ fuzzy. The many material variations plus nonuniformity in construction and processing can significantly alter $\mathrm{S}_{\mathrm{d}}$. But, you can at least calculate a value for $S_{d}$ that is close enough for a first-order estimate.

Find $\mathrm{S}_{\mathrm{d}}$ at the ambient temperature, $\mathrm{T}_{\mathrm{a}}$, by

$$
\begin{equation*}
\mathrm{S}_{\mathrm{d}}=0.0014+1.217 \times 10^{-6} \mathrm{~T}_{\mathrm{a}}{ }^{1.585} \tag{34}
\end{equation*}
$$

With the total surface area, $A_{s}$, and $\mathrm{S}_{\mathrm{d}}$, approximate thermal conductance:

$$
\begin{equation*}
\mathrm{G}_{\mathrm{t}}=\mathrm{A}_{\mathrm{s}} \mathrm{~S}_{\mathrm{d}}, \tag{35}
\end{equation*}
$$

Then the temperature rise:

$$
\begin{equation*}
\mathrm{T}_{\mathrm{r}}=\mathrm{P}_{\mathrm{t}} / \mathrm{G}_{\mathrm{t}}, \tag{36}
\end{equation*}
$$

and finally the crucial variable, $\mathrm{T}_{\mathrm{m}}$ :

$$
\begin{equation*}
\mathrm{T}_{\mathrm{m}}=\mathrm{T}_{\mathrm{a}}+\mathrm{T}_{\mathrm{r}} . \tag{37}
\end{equation*}
$$

If you find the $T_{m}$ gets too high, reduce it by increasing $A_{s}, R_{d c}$, or both.

## Putting the steps all together

To see how all the equations work together, consider an inductor meeting the following requirements:

- $\mathrm{L}_{0}=5.6 \mathrm{mH}$ at 0 -Adc.
- $\mathrm{L}_{\mathrm{m}}=4.7 \mathrm{mH}$ minimum at $1.4-\mathrm{A} \mathrm{pk}$ and 20 kHz with a $34 \%$ duty cycle.
- $\mathrm{T}_{\mathrm{r}}=60 \mathrm{C}$ maximum at 20 C ambient, therefore $\mathrm{T}_{\mathrm{m}}=80 \mathrm{C}$.
- $\mathrm{I}_{\mathrm{dc}}=0$.
- The inductor must be mountable on a PC board and shielded.
- The inductor must be small.

Step 1: Determine the total current. Because $\mathrm{I}_{\mathrm{dc}}$ is zero, from Eq. 1,

$$
\begin{aligned}
\mathrm{I}_{\mathrm{m}} & =1.4+0 \\
& =1.4 \mathrm{~A}
\end{aligned}
$$

For a $34 \%$ duty cycle at 20 kHz ,

$$
\begin{aligned}
\mathrm{t}_{\mathrm{pr}} & =1 / 20 \times 10^{3} \\
& =5 \times 10^{-5} \mathrm{~s}, \\
\mathrm{P}_{\mathrm{d}} & =0.34 \times 5 \times 10^{-5} \\
& =1.7 \times 10^{-5} \mathrm{~s} .
\end{aligned}
$$

and
Assuming that $\mathrm{I}_{\mathrm{c}}$ is $10 \%$ of $\mathrm{I}_{\mathrm{m}}$,

$$
\begin{aligned}
\mathrm{I}_{\mathrm{c}} & =0.1 \times 1.4 \\
& =0.14 \mathrm{~A} .
\end{aligned}
$$

Although not specified in this design, $\mathrm{E}_{\mathrm{pk}}$ is always of interest, so from Eq. 5,

$$
\begin{aligned}
\mathrm{E}_{\mathrm{pk}} & =(1.4-0.14) 4.7 \times 10^{-3} /\left(1.7 \times 10^{-5}\right) \\
& =348 \mathrm{~V} \mathrm{pk} .
\end{aligned}
$$

From Eq. 6,

$$
I_{e f f}=\sqrt{\frac{1.7 \times 10^{-5}\left[1.4^{2}+(1.4 \times 0.14)+0.14^{2}\right]}{3 \times 5 \times 10^{-5}}}
$$

And from Eq. 2,

$$
\begin{aligned}
\mathrm{I}_{\mathrm{r}} & =0.497+0 \\
& =0.497 \mathrm{~A} .
\end{aligned}
$$

Step 2: Select the core material. Choose F2000 ferrite for the core material because, as you can see from Table 1, it has the greatest operating flux density. From the table,

$$
\begin{aligned}
& \mathrm{B}_{\mathrm{s}}=4400 \mathrm{G} \\
& \mathrm{~B}_{\mathrm{r}}=1500 \mathrm{G} .
\end{aligned}
$$

And from Table 2,

$$
\begin{aligned}
\mathrm{U}_{\mathrm{i}} & =2000 \mathrm{G} . \\
\mathrm{T}_{\mathrm{m}} & =+125 \mathrm{C} .
\end{aligned}
$$

From the supplier's data, at 2900 G and 20 kHz ,

$$
\mathrm{C}_{\mathrm{vf}}=12 \mu \mathrm{~W} / \mathrm{cm}^{3} / \mathrm{Hz} .
$$

Therefore, from Eq. 7,

$$
\begin{aligned}
\mathrm{C}_{\mathrm{v}} & =12 \times 10^{-6} \times 20 \times 10^{3} \\
& =0.24 \mathrm{~W} / \mathrm{cm}^{3}
\end{aligned}
$$

Step 3: Determine the optimum B and H. From Eq. 9,

$$
\begin{aligned}
\mathrm{B} & =4400-1500 \\
& =2900 \mathrm{G} .
\end{aligned}
$$

From Eq. 10,

$$
\begin{aligned}
\mathrm{U}_{\mathrm{m}} & =2000 \times 4.7 \times 10^{-3} / 5.6 \times 10^{-3} \\
& =1679 \mathrm{G} / 0 \mathrm{e} .
\end{aligned}
$$

From Eq. 11,

$$
\begin{aligned}
\mathrm{H} & =2900 / 1679 \\
& =1.730 \mathrm{e} .
\end{aligned}
$$

Step 4: Select the minimum-sized magnet wire. With an $80-\mathrm{C}$ value for $\mathrm{T}_{\mathrm{m}}$ in Eq. 14,

$$
\begin{aligned}
\mathrm{I}_{\mathrm{ro}} & =0.497[1+0.00393(80-20)] \\
& =0.614 \mathrm{~A} .
\end{aligned}
$$

From the data in Table 3, the thinnest wire that can accommodate $\mathrm{I}_{\mathrm{ro}}$ is AWG 33, whose

$$
\mathrm{p}=0.206 \Omega / \mathrm{ft},
$$

and

$$
\mathrm{d}_{\mathrm{i}}=0.0088 \mathrm{in} .
$$

Step 5: Determine the minimum effective volume. From Eq. 15,

$$
\begin{aligned}
\mathrm{V}_{\mathrm{em}} & =0.4 \pi \times 10^{8} \times 4.7 \times 10^{-3} \times 1.4^{2} /(2900 \times 1.73) \\
& =231 \mathrm{~cm}^{3}
\end{aligned}
$$

To provide this effective volume, an upgapped core would have to fill a $6.13-\mathrm{cm}$ cube-much too large for mounting on a PC board. Use a gapped core instead; and since shielding is required, make it a pot core.
Step 6: Select the minimum-sized core. From Eq. 18,
$\mathrm{W}_{\mathrm{a}} \mathrm{A}_{\mathrm{c}}=5.067 \times 10^{8} \times 4.7 \times 10^{-3} \times 1.4 \times 0.0088^{2}$
$0.8 \times 2900$

$$
=0.1113 \mathrm{~cm}^{4} .
$$

A search through core catalogs reveals the smallest standard pot core that you can use is $22 \times 13 \mathrm{~mm}$. For this core

$$
\begin{aligned}
\mathrm{A}_{\mathrm{c}} & =0.63 \mathrm{~cm}^{2} \\
\mathrm{~W}_{\mathrm{a}} & =0.292 \mathrm{~cm}^{2} . \\
\mathrm{W}_{\mathrm{a}} \mathrm{~A}_{\mathrm{c}} & =0.292 \times 0.63 \\
& =0.1840 \mathrm{~cm}^{4},
\end{aligned}
$$

and
Therefore,
which is, of course, large enough. From the catalogs, you get the following additional core data:

$$
\begin{aligned}
\mathrm{l}_{\mathrm{c}} & =3.15 \mathrm{~cm} . \\
\mathrm{V}_{\mathrm{c}} & =2 \mathrm{~cm} . \\
\mathrm{A}_{\mathrm{s}} & =18.02 \mathrm{~cm}^{2} . \\
\mathrm{W}_{\mathrm{o}} & =0.0453 \mathrm{in} . \\
\mathrm{l}_{\mathrm{tm}} & =0.145 \mathrm{ft} .
\end{aligned}
$$

Step 7: Determine the number of turns. From Eq. 19,

$$
\begin{aligned}
\mathrm{A}_{\mathrm{Lm}} & =(2900 \times 0.63)^{2} \times 10^{-16} /\left(4.7 \times 10^{-3} \times 1.4^{2}\right) \\
& =3.623 \times 10^{-8} \mathrm{H} / \text { turns }^{2} .
\end{aligned}
$$

Then from Eq. 20

$$
\begin{aligned}
\mathrm{N} & =\sqrt{4.7 \times 10^{-3} /\left(3.623 \times 10^{-8}\right)} \\
& =360 \text { turns }
\end{aligned}
$$

Step 8: Determine the air gap size. From Eq. 21,

$$
\begin{aligned}
\mathrm{l}_{\mathrm{e}} & =0.4 \pi \times 360 \times 1.4 / 1.73 \\
& =366 \mathrm{~cm}
\end{aligned}
$$

Then, from Eq. 23,

$$
\begin{aligned}
\mathrm{l} \mathrm{~g} & =(366-3.15) / 2000 \\
& =0.182 \mathrm{~cm} \\
& =0.072 \mathrm{in} .
\end{aligned}
$$

The air gap interrupts the magnetic circuit twice, so the spacer should be 0.036 in. thick-half the computed value of $\mathrm{l}_{\mathrm{g}}$.

Check $\mathrm{V}_{\mathrm{e}}$ in Eq. 16:

$$
\begin{aligned}
\mathrm{V}_{\mathrm{e}} & =2+(0.63 \times 2000 \times 0.182) \\
& =231 \mathrm{~cm}^{3}
\end{aligned}
$$

Therefore, $\mathrm{V}_{\mathrm{e}}$ complies with the $\mathrm{V}_{\mathrm{em}}$ required by Eq. 17.

Step 9: Determine the incremental effect. From Eq. 25 ,

$$
\begin{aligned}
\mathrm{A}_{\mathrm{LO}} & =0.4 \pi \times 10^{-8} \times 0.63 /[(3.15 / 2000)+0.182] \\
& =4.31 \times 10^{-8} \mathrm{H} / \text { turn } .
\end{aligned}
$$

Then, from Eq. 26,

$$
\begin{aligned}
\mathrm{L}_{\mathrm{o}} & =4.31 \times 10^{-8} \times 360^{2} \\
& =5.59 \mathrm{mH},
\end{aligned}
$$

which is close enough to the required 5.6 mH . If $L_{0}$ isn't close enough, change the turns ratio ap-propriately-a simple process using Eq. 26.

Step 10: Determine the total power loss. From Eq. 29,

$$
\begin{aligned}
\text { tpi }^{2} & =360 /(0.8 \times 0.0453) \\
& =9934 \text { turns } / \text { in. }^{2}
\end{aligned}
$$

From the wire data in Table 2, you can use AWG 32 heavy-film wire instead of AWG 33. Then,

$$
\mathrm{p}=0.162 \Omega / \mathrm{ft} .
$$

From Eq. 30,

$$
\begin{aligned}
\mathrm{R}_{\mathrm{dc}} & =360 \times 0.145 \times 0.162 \\
& =8.46
\end{aligned}
$$

From Eq. 31,

$$
\begin{aligned}
\mathrm{P}_{\mathrm{w}} & =0.497^{2} \times 8.46[1+0.00393(80-20)] \\
& =2.54 \mathrm{~W} .
\end{aligned}
$$

From Eq. 32,

$$
\begin{aligned}
P_{c} & =0.24 \times 2 \\
& =0.48 \mathrm{~W} .
\end{aligned}
$$

Note that the core loss is much smaller than the winding loss. From Eq. 33,

$$
\begin{aligned}
\mathrm{P}_{\mathrm{t}} & =0.48+2.54 \\
& =3.02 \mathrm{~W} .
\end{aligned}
$$

Step 11: Determine the maximum temperature. From Eq. 34,

$$
\begin{aligned}
\mathrm{S}_{\mathrm{d}} & =0.0014+1.2717 \times 10^{-6} \times 25^{1.585} \\
& =0.0016 \mathrm{~W} / \mathrm{cm}^{2} /{ }^{\circ} \mathrm{C} .
\end{aligned}
$$

From Eq. 35,

$$
\begin{aligned}
\mathrm{G}_{\mathrm{t}} & =0.0016 \times 18.02 \\
& =0.02883 \mathrm{~W} /{ }^{\circ} \mathrm{C} .
\end{aligned}
$$

From Eq. 36,

$$
\begin{aligned}
\mathrm{T}_{\mathrm{r}} & =3.02 / 0.02883 \\
& =105 \mathrm{C} .
\end{aligned}
$$

From Eq. 37,

$$
\begin{aligned}
\mathrm{T}_{\mathrm{m}} & =20+105 \\
& =125 \mathrm{C} .
\end{aligned}
$$

This value of $T_{m}$ is too high. The open construction of the proposed inductor doesn't have enough surface to limit $\mathrm{T}_{\mathrm{r}}$ to 60 C max. From Eq. 36, the minimum thermal conductance required is

$$
\begin{aligned}
\mathrm{G}_{\mathrm{t}} & =3.02 / 60 \\
& =0.0503 \mathrm{~W} /{ }^{\circ} \mathrm{C}
\end{aligned}
$$

Therefore, from Eq. 35, the minimum $\mathrm{A}_{\mathrm{s}}$ required is

$$
\begin{aligned}
\mathrm{A}_{\mathrm{s}} & =0.0503 / 0.0016 \\
& =31.43 \mathrm{~cm}^{2} .
\end{aligned}
$$

The maximum diameter and height of the pot core are 0.866 inches and 0.536 inches, respectively. So you should be able to encapsulate it, with thermally conductive epoxy, in a round or rectangular plastic shell. The nearest suitably sized round shell has an outside diameter of 1.187 inches, a height of 0.75 inches, and a thickness of 0.03 inches. Therefore,

$$
\begin{aligned}
\mathrm{A}_{\mathrm{s}} & =5.01 \mathrm{in} .^{2} \\
& =32.32 \mathrm{~cm}^{2} .
\end{aligned}
$$

Combining Eqs. 35, 36 and 37 ,

$$
\begin{aligned}
\mathrm{T}_{\mathrm{m}} & =20+3.02 / 0.0016 \times 32.32 \\
& =20+58 \\
& =78 \mathrm{C}
\end{aligned}
$$

which meets the required design goal. $\quad$ -

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## SWTITHER OUTPUT FIITER CAPACTTORS That Really PUT OUT

If you're working with switchingtype power supplies, you'll want to know about new electrolytic capacitors featuring low equivalent series resistance (for example, 3 ms @ $57,000 \mu \mathrm{~F} / 7.5 \mathrm{~V}$ ) and low internal inductance. Type 622D EXTRALYTIC ${ }^{\wedge}$ Capacitors are the first of their type to meet the power supply designer's need to know, for worst case design, the maximum and minimum ESR of a capacitor, as well as the need to hold the nominal ESR to a tolerance of $\pm 30 \%$ at 20 kHz . To simplify calculations for the

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For complete technical data, write for Engineering Bulletin 3459 to: Technical Literature Service, Sprague Electric Company, 347 Marshall Street, North Adams, Mass. 01247.

THE BROAD-LINE PRODUCER OF ELECTRONIC PARTS

Technology

# Consider mask-programmable arrays instead of custom-designed circuits. Arrays offer the flexibility of custom circuits with almost 'stock' delivery times. 

When the design you're working on calls for a custom circuit, but you can't tolerate a six to 12-month design turnaround or a large cost, try an uncommitted array and get a custom part for almost standard prices. Available with either digital or linear formats, an uncommitted array becomes a price and performance alternative to a custom circuit since just a final metallization layer must be deposited to interconnect the prediffused transistors, gates, and passive components.

Developing a "from the ground up" custom design can cost about $\$ 50,000$ on up, depending on the complexity of the IC. Development and processing can take anywhere from six months to well over a year. With a mask-programmable array, however, your development time can be cut to several weeks-the time it takes to generate the metal masks from an interconnection diagram. Moreover, you can keep development to about $\$ 10,000$-and get prototypes in six to eight weeks.

Of course, the development dollars that an array vendor puts into an uncommitted device array must be repaid somehow, so you will end up paying his costs in the form of higher chip prices than those charged by a custom IC manufacturer. The custom IC vendor does all the development from the ground up, thus shaving chip costs to the bone by keeping the silicon area of the die to a minimum.
But you can't really compare total design costs without knowing how many devices using arrays will be produced. In fact since production cost per unit for a purely custom device will drop more rapidly than the cost of an uncommitted array, any economic decision you make depends almost entirely on the number of chips to be purchased.
For 10,000 and under, uncommitted arrays are usually more favorable, while for more than 10,000 a totally custom circuit usually looks more attractive. Table 1 shows how the two alternatives compare when 10,000 units are purchased yearly, and the development costs for a similar circuit are estimated at $\$ 50,000$ and $\$ 10,000$ for the custom and uncommitted circuits, respectively.
Although uncommitted arrays look good for low

[^3]

Typical of uncommitted digital arrays, this MasterMOS Model S chip contains 106 p-channel and 106 n-channel transistors, 10 medium-power buffers, four large n-channel devices with high-current capability, and one p-channel and one n-channel high-impedance device. The chip is $65 \times 74$ mils, with pads of $4 \times 4$ mils.
volumes, less than a dozen companies offer maskprogrammable units (Table 2). The differences in these arrays are essentially the differences between the semiconductor processes they are made from. For example, if you're looking for a CMOS array, you can go only to about three companies-and you will find little, if any, differences between the arrays.

## Check the process and performance

How many processing steps are required to produce finished devices also depends on technology. CMOS wafers require only one step-etching the aluminum interconnects-while other technologies require two or more steps.
Of course, you'll have to make some performance compromises between the mask-programmed arrays and custom-designed circuits. CMOS arrays, for in-
stance, operate in the 1 to $4-\mathrm{MHz}$ region and require several microwatts of power. True custom devices typically can perform at slightly higher frequencies and at lower power levels.
In addition, since most available arrays are pointed toward digital applications, most wafers are made with many small circuit elements or gates. The small elements are not suited for linear applications, and another array must often be used when designing linear devices (some typical array specs for the CMOS digital version are listed in Table 3).

Still, designing with an uncommitted array is fairly simple, since all the transistors are already placed on the wafer. All you have to worry about is designing the logic of the functions to be implemented on the
Table 1. Custom cost vs array cost

|  | Custom | Prediffused |
| :--- | :--- | :--- |
| Development cost | $\$ 50,000$ | $\$ 10,000$ |
|  | $(\$ 5 /$ unit $)$ | $(\$ 1 / \mathrm{unit})$ |
| Production cost | $\$ 20,000$ | $\$ 30,000$ |
|  | (\$2/unit) | (\$3/unit) |
| Total | $\$ 70,000$ | $\$ 40,000$ |
|  | (\$7/unit) | (\$4/unit) |

chip. Then, working with an array manufacturer, you figure out the interconnections needed on the chip.

## Divide the procedure into steps

Although the design procedure will vary slightly from one company to the next, depending on the technology, the CMOS arrays from International Microcircuits can be designed by following six simple sequential steps:

1. Define circuit functions and design the logic required to implement those functions. Breadboarding is often valuable at this point for checking the circuit functions and verifying all operating modes.
2. Convert the detailed logic design to a complete circuit design that includes every transistor and every logic element. You should now count logic elements and input and output lines to help you select the rightsized array for your design.
3. Sketch interconnecting lines on a large drawing of the appropriate array. This requires part intuition and part trial and error to arrive at the best grouping of components and eliminate any crossovers. The sketch need only show the points that are to be connected, and not the exact routing of the interconnecting lines.

The three previous steps can be done without the

## Table 2. Major uncommitted array vendors

| Manufacturer | Array capability | Circle No. |
| :---: | :---: | :---: |
| Exar, 750 Palomar Ave., Sunnyvale, CA 94086. (408) 833-7700. | Bipolar (linear and $1^{2} \mathrm{~L}$ digital) | 511 |
| Ferranti, E. Bethpage Rd., Plainview, NY 11803. (516) 293-8383. | Bipolar | 512 |
| Interdesign, 1255 Reamwood Ave., Sunnyvale, CA 94086. (408) $734-8666$. | Bipolar, CMOS and NMOS | 513 |
| International Microcircuits, 3004 Lawrence Expressway, Santa Clara, CA 95051. (408) 735-9370. | CMOS | 514 |
| Master Logic, 761 E. Evelyn Ave., Sunnyvale, CA 94086. (408) 732-7777. | CMOS | 515 |
| Microcircuits Technology, 975 Comstock St., Santa Clara, CA 95050. (408) 249-2501. | CMOS, NMOS and PMOS | 516 |
| Motorola Semiconductor, 5005 E. McDowell Rd., Phoenix, AZ 85008. (602) 244-6900. | Bipolar (ECL) | 517 |
| RCA Solid State, Route 202, Somerville, NJ 08876. (201) 685-6000. | CMOS | 518 |
| Siemens, Oskar-von-Miller Ring 18, D-8000, Munchen 2, Federal Republic of Germany. | Bipolar (ECL) | 519 |
| Stewart Warner Microcircuits SWAP (now part of Dionics, 65 Rushmore St., Westbury, NY 11590. (516) 997-7474). | Bipolar ( ${ }^{2} \mathrm{~L}$ ) | 520 |
| TRW, 14520 Aviation Blvd., Lawndale, CA 90260. (213) 679-4561. | Bipolar (TTL) | 521 |




Designing a circuit with an uncommitted array can be broken into three steps. Start with the logic diagram of the circuit (left), which in this example is a four-channel scanner, and then transform the logic diagram into a discrete device diagram (above) so that the metal interconnect patterns can be defined (below).


Table 3. Sample array specifications


- assumes a D-type flip-flop built from array devices
aid of the array vendor, but the next three steps usually require hand-in-hand cooperation:

4. Working from the sketch, technicians prepare precise artwork that indicates all interconnecting lines. The artwork is then checked against the logic design and circuit design to minimize the possibility of a layout error.
5. A photomask is produced from the artwork and is used to etch the aluminum interconnections on the preprocessed wafer.
6. The finished chips are inspected, diced, packaged, tested and then shipped to the customer.

The array manufacturer can enter this design schedule at any time up to the preparation of the photomask artwork. However, the earlier the vendor gets involved, the easier the design procedure. Some vendors even maintain a staff of applications engineers that can help you design your circuit, or at least give you some basic guidelines as to what the arrays can and can't do...


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# Four-phase logic is practical when implemented with high-density MOS large-scale arrays. Low power and high speed can overshadow any disadvantages. 

The high packing densities of MOS circuits now allow designers of digital systems to take full advantage of four-phase logic economically. Though fourphase logic is not a new concept, ${ }^{1}$ it isn't widely understood, so here's a quick review of its characteristics.

Four-phase logic offers the following advantages:

- Very low power consumption-only a few microwatts per stage.
- Higher speeds than conventional DTL, TTL and even many CMOS-logic systems-typically only $10-\mathrm{ns}$ delay in a four-phase stage.
- Simple circuitry, highly compatible with computer-aided methods-all "components" are MOS devices.
But four-phase logic also has some disadvantages:
- The logic system is dynamic-inputs and outputs occur only in specific time slots.
- The system needs a four-phase clock-the timing intervals must be accurately determined.
- The rules of interconnection must be strictly observed-even though they are simple and easily handled by computer-aided design.
Nevertheless, the advantages outweigh the disadvantages, especially in large-scale programmablelogic arrays (PLAs), programmable-gate arrays, and $\mu$ Ps. And converting conventional logic circuits into four-phase logic is simple-once you know how fourphase logic works.


## Inherent capacitance holds the signal

Fig. 1 shows a basic four-phase "stage" configured as an inverter. Phase-1 ( $\varnothing 1$ ) of the system's clock powers the series string of MOSFET transistors $T_{1}$, $\mathrm{T}_{2}$ and $\mathrm{T}_{3}$; phase $2(\varnothing 2)$ clocks the gate input of $\mathrm{T}_{2}$; and the logic-signal input is applied to the gate of $\mathrm{T}_{3}$.
The $\varnothing 1$ clock, applied to the gate and drain of $\mathrm{T}_{1}$, turns $\mathrm{T}_{1}$ on, and negatively charges the small inherent capacitance, C , of the next stage's input gate. After this charging time (CT), $\varnothing 2$, which holds $\mathrm{T}_{2}$ on, allows the charge on C to remain or reverse, depending on the logic-signal input to $\mathrm{T}_{3}$.
A ONE (corresponding to a negative level) on the

[^4]

1. In a four-phase inverter, the inherent small capacitance of the next stage's input charges during time CT of clock $\varnothing 1$. The state of this capacitance is determined, during the input time (IT), by the condition of the logic input A. During the output time (OT), the output-the state of $C$-transfers to the next stage.

2. Four-phase logic requires the use of four stage configurations. The rules for stage interconnection are relatively simple. Besides, the different stage delays add logic flexibility, and available computer programs can easily obey all the rules consistently.

3. A four-phase clock is needed for four-phase logic (a), and the clock-signal transitions define the output time (OT), input time (IT), charge time (CT) and dummy, or idle, time (DT) of each of the four stage configurations.
input to $T_{1}$ turns $T_{3}$ on, discharges $C$ and recharges it positively to ZERO, since $ø 1$ is now positive and $\mathrm{T}_{1}$ has turned off. Conversely, a ZERO (positive level) input to $T_{3}$ retains the negative charge on $C$; clearly, the stage acts as an inverter. This part of the circuit's operation is called the input time (IT).

When $\phi 1$ and $\phi 2$ are both positive, $\mathrm{T}_{1}$ and $\mathrm{T}_{2}$ are cut off and C is isolated. This part of the operation is called the output time (OT). Another operational interval, called idle-or dummy-time (DT), doesn't occur in the stage type just discussed-Stage 1-but does in two other types of stages-Stages 2 and 4 that we'll look at later (Fig. 2).
The voltage polarity that represents a ONE or ZERO depends upon the logic fabrication-PMOS or NMOS. Since PMOS gates conduct when the gate is negative with respect to the source, the convention is to use ONEs to represent negative levels.

Low-threshold PMOS four-phase logic systems readily operate with $7-$ to- $8-\mathrm{MHz}$ clocks, and well
designed NMOS systems can work up to 16 MHz . For example, General Instruments' MEM 3064, a 64 -bit serial accumulator, operates at 5 MHz , contains 400 PMOS devices on a $58 \times 58$-mil chip, and dissipates only 40 mW . And LSI Computer Co. (Melville, NY) is developing four-phase custom PMOS and NMOS arrays for miniaturized computers like the Control Data 469 miniature computer.

## Four clock phases are needed

Fig. 3a shows the complete set of timing and wave shapes-the four phases of the clock needed to power a complete PMOS four-phase logic system (other wave configurations are possible ${ }^{1}$ ). ONEs and ZEROs propagate-as in a shift register-through a logic system made up of properly interconnected combinations of the four coupling, or "stage," circuits shown in Fig. 2. Clock pulses under control of the logic conditions charge and discharge the small inherent capacitances of the logic-input gates of each successive stage. The gate capacitance is, about 0.2 pF typically, with a series impedance of about $25 \mathrm{k} \Omega$; therefore, a stage output can easily drive a dozen other stages

In addition to four clock phases, four stage configurdtions are needed (Fig. 2) because, unfortunately, a stage can't drive another with the same configuration; nor can the stages be connected to each other randomly. A stage may drive its next consecutively numbered stage-Stage 1 may drive Stage 2, 2 may drive 3, and so on. Stage 4 then drives Stage 1. In addition, Stages 1 and 3 may drive each other, so there are six legal combinations in all. All other stage connections are illegal-for example, Stages 2 and 4 can't connect to each other directly.

Furthermore, the logic-signal inputs to Stages 2 and 4 don't propagate through a clocked MOS gate to get to the output; thus their outputs are immediate (real time), whereas stages 1 and 3 outputs are delayed half a clock time of $\phi 2$ or $\phi 4$, respectively. Also, note that the outputs of all the stage types are inverted.

These legal-connection rules may seem complicated, at first. But they are easy to remember and they give you considerable flexibility, especially when you design dynamic and sequential systems, such as shifting, counting and time multiplexing circuits. For example, a Stage-1 inverter connected to a Stage-3 inverter generates an output equal to the input, but delayed by one bit. This combination of stages, therefore,
IMMEDIATE
4. Logic functions implemented in four-phase logic must be "carried" in one of the four stage configurations as
governed by the interconnection rules. The logic function in a stage can be simple or complex, as desired.
becomes one stage of a shift register.
Fig. 4 shows a few examples of some simple logic functions in four-phase configurations: an inverter, a NAND gate, a NOR gate and two NAND/NOR combinational circuits. You can easily compile a collection of your own simple-logic building blocks.

But, if you're building complex logic structures, take care: Often the MOS characteristics limit the circuit to only two or three serial OR-gate logic functions. To overcome serial-OR limitations, alternate verticalparallel OR (Fig. 4c) and horizontal-parallel-OR (Fig. 4d) configurations. The difference between the vertical and parallel configurations lies in the way the IC-chip designer structures the gates and current paths in the MOS devices. Further, to optimize the circuit density on the chip, the use of real-time stages and delayed stages should be alternated.

## Five steps to four-phase logic

Designing four-phase logic circuits boils down to five simple steps. Note that the first four steps are merely the steps of ordinary logic design:

1. Draw a truth table.
2. Map the truth table on a Karnaugh map.
3. Write the logic equation in its simplest form.
4. Lay out the circuit in conventional logic and use one type of logic gate only-all ANDs, ORs, NANDs or NORs.
5. Transform the design into a four-phase circuit, drawing the MOS gates as bubbles.

A conventional logic design can be transformed into four-phase logic many ways-the number is directly proportional to its complexity. For example, Fig. 5 takes you through the steps in designing a two-bit full
adder. However, Step 4, which calls for a conventional design, can be skipped, because of the circuit's simplicity.

Fig. 5 c is a straightforward implementation for the sum, S, of the equation in Fig. 5b. However, the application of De Morgan's theorem, ${ }^{2}$

$$
\begin{aligned}
& A \cdot B \cdot C \ldots N=\bar{A}+\overline{\mathrm{B}}+\overline{\mathrm{C}} \cdot \overline{\mathrm{~N}}_{1} \\
& \mathrm{~A}+\mathrm{B}+\mathrm{C}+\ldots \mathrm{N}=\overline{\mathrm{A}} \cdot \overline{\mathrm{~B}} \cdot \overline{\mathrm{C}} \cdot \ldots
\end{aligned}
$$

eliminates the need for a separate inverter after each logic term before the ORing operation, and thus greatly simplifies the logic design in Fig. 5d.
The logic for the carry, $\mathrm{C}_{(\mathrm{n}+1)}$, also is shown in two versions: Fig. 5 f is a straightforward application of the equation in Fig. 5b, and Fig. $5 f$ is a somewhat simpler version..п

## References

1. Boysel L.L., and Murphy, J.P., "Four-Phase LSI Logic Offers New Approach to Computer Designer," Computer Design, April, 1970, pp. 141-146.
2. Asija, S.P., "Instant Logic Conversion," IEEE Spectrum, December, 1968, pp. 77-80.


3. A two-bit full-adder design example is taken through the suggested design steps from truth table (a) and Karnaugh map (b) to variations of the design-direct
implementation (c and e)—or use of De Morgan's theorem (d and f). However, because of the simplicity. Step 4the conventional-logic design-is skipped.


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# Shrinkable plastic tubes, boots and caps can solve many of your insulating problems neatly and efficiently. Even irregular shapes can be fully protected. 

Heat shrinkable plastics are a substantial portion of the insulating materials used today. Tubing is by far the commonest form, but boots, caps and transition pieces also are used in large quantities. Even many irregularly shaped objects, formerly insulated with tape, are now protected with special heatshrinkable forms rapidly and at low cost.

The good insulating and mechanical properties of most shrinkable plastic tubing, along with its compatibility with ordinary wire-insulation plastics make such tubing eminently suitable for covering splices to provide both electrical and mechanical protection. As an added bonus, splices have a neater appearance with tubing, rather than with electrical tape.

Furthermore, shrinkable tubing is easy and convenient to use. It cuts easily with scissors, electrical cutters or standard paper cutters. And automated cutting equipment is available. Some tubing manufacturers and distributors even supply pieces cut to order. Standard industry packages contain $4-\mathrm{ft}$ tubing lengths, but some suppliers offer 2 -ft lengths, which are easier to store and handle. But even 6 -in. packages are readily available.

You can fabricate your own cables, and jacket them for both protection and identification. Damaged flexible cord and cable jackets can be conveniently repaired with shrinkable sleeving, and thus protected against any further damage or moisture penetration. Encapsulating and moisture-proof tubing can seal connectors and splices. Also, shrinkable tubing can serve as strain-relief bushing for cords and connectors. And besides being available in a wide range of colors, the tubing can be surface-printed with characters.

Another major tubing application is to insulate connector terminations. The tubing is slipped over individual connector wires before they are terminated. After termination, the tubing is placed over the wire/connector-terminal joint and shrunk to both insulate the connection and relieve it from strain.
In addition, end caps, which are tubes closed on one end, are used to protect bundles of splices and to prevent moisture from penetrating the cut ends of cable when in storage. And connector boots are specially shaped to fit connector profiles. They protect

[^5]individual wire terminations, provide strain relief for the connector and form an uninterrupted covering from the original cable jacket to the connector entrance into a panel or equipment.
Several different polymeric thermoplastic materials can be made shrinkable, but the basic molecular interactions that produce this shrinkable behavior are common to all of them. The key to shrinkable behavior lies in strengthening the links among the long-chain molecules within a polymer's crystal structure. One way to strengthen these links is to expose the material to radiation under controlled conditions. The radiation produces a crosslinking of the molecules that strongly binds the long-chains together.
With crosslinking, if the material is heated to the crystalline melting point, it doesn't melt, but becomes soft and rubbery. The crystal structure dissolves, but the crosslinks hold the molecules together in a viscous mass that prevents the material from melting and flowing. In this soft state, the material can be stretched, which puts the crosslinks under tension. If cooled while held in the expanded state, the crystal structure reforms, but the material retains its expanded dimensions indefinitely, until reheated.
When reheated to the crystalline melting points, the crystal structure disappears again, the crosslink tension relaxes and the material returns to its preexpanded dimensions. When cooled, the material has come full circle to provide the user with a tough material that fits snugly on the covered component.

## Pick the right material

The large choice of available shrinkable materials may make selection difficult. But at the same time, the large range of properties allows you to meet your needs more precisely. The following short list of the most commonly used materials should help you find what you want:

1. Irradiated flexible polyolefin is by far the most commonly used shrinkable plastic. The material has good electrical and mechanical properties for most general-purpose applications. It is popular because it is both low-cost and available in a wide range of sizes and colors. Shrink temperature is a cool 121 C .
2. Irradiated semirigid polyolefin is a physically stiffer and stronger version of polyolefin, and it's used

3. Heat-shrinkable tubing is available in many sizes and types of materials to suit almost any application.
primarily for the strain relief of soldered connections and terminals. Shrink temperature is 135 C and price is about $1-1 / 3$ times the flexible variety.
4. Surface-irradiated polyolefin behaves differently from the other shrinkable materials. Because only its outer surface is irradiated and shrinkable, the inner surface can melt and flow when heated to the shrinking temperature. When the outer surface shrinks, the melted inner material squeezes into the crevices of the covered item to provide excellent mechanical adherence and protection against vibration and strain. Shrink temperature is 135 C , and its price is about three times that of flexible polyolefin.
5. Mastic-lined polyolefin is coated inside with meltable mastic. As the tubing is heated and shrinks, the mastic melts and flows. Unlike the interior of surface-irradiated polyolefin, which hardens completely when cooled, the mastic material remains somewhat viscous to provide a strong barrier against moisture. The material is particularly suitable for sealing outdoor connections and for protecting connec-

6. Heating devices for shrinkable tubing come in many styles and with a variety of attachments to provide uniform and rapid heating where needed.
tions directly buried in the ground. Shrink temperature is 135 C , but the price is high-about $5-1 / 2$ times flexible polyolefin.
7. Irradiated vinyl is the most economical of all the shrinkable tubing. Unlike most of the other materials, it is available in long lengths, so it is suitable for jacketing long cables. However, vinyl's insulating capability is slightly inferior to polyolefin, and is available only in black. Shrink temperature is high -175 C . A nonirradiated shrinkable vinyl is also available, but the irradiated material offers several distinct advantages: it has indefinite shelf life, withstands solder-iron heat, resists splitting and possesses low longitudinal shrinkage.
8. Irradiated Kynar is a high-temperature plastic capable of continuous operation at 175 C . It is nonflammable, resistant to severe environmental conditions and physically very strong. Because of its high strength, Kynar tubing can be very thin walled. Standard tubing is transparent and shrink temperature is 175 C . The material is about three times more expensive than flexible polyolefin.
9. Irradiated neoprene, also available in long tube lengths, is very flexible and durable as a cable jacket, but it is used more for mechanical protection than for electrical insulation. It comes in black and its shrink temperature is 175 C . Price is about five times that of flexible polyolefin.
10. Teflon, or tetrafluoroethylene in chemical terminology, doesn't melt and therefore doesn't require irradiation. Instead, Teflon undergoes a crystal-structure transition at 327 C , which makes it pliable so that it can be stretched. The molecular bonds, thus put under tension, when reheated to 327 C are relieved and the material shrinks. Since Teflon doesn't melt or burn, and resists almost all chemical attack, it can fit many tough applications. But its price is over 4-1/2 times that of flexible polyolefin.
11. Fluorinated ethylene propylene, of FEP, is a meltable grade of Teflon. Shrinkable FEP has the chemical-resistance properties of TFE, but not the continuous-duty rating-204 C as opposed to 250 C

12. The hot-air gun is the most popular heating tool for shrinkable material. It can be used to shrink a feedthrough seal around a heavy-duty control cable (left) or
for TFE. It's easier to install, however, because its shrink temperature is a lower 176 C . But it costs nearly five times more than flexible polyolefin.

## Shrinking is easy

Shrinkable materials can be heated with a cigarette lighter or even a match. But large tubing sizes can't be heated evenly with a small flame. Also, open flames often can't be used in some environments, sensitive components can be damaged and soot residues can destroy the material's appearance and compromise its insulation protection. A hot-air gun, however, is both safe and clean.

A gun can supply a required shrink temperature uniformly over a large area. An electric heating element raises the temperature of air, which is forced out of the gun nozzle with a fan. Guns come in many sizes and for different temperatures. In some, the temperature can be precisely adjusted. And by using differently shaped nozzles, an operator can concentrate the gun's output as required. The results obtained will be uniform and quick. Obviously, shrinking should take as short a time as possible, both to maximize production and to protect heat-sensitive circuit components.
But electric power isn't always available, say, to seal pipe joints and to splice large underground power and communications cables. Very large tubing used outdoors is often shrunk with a propane torch. At other times, for delicate small jobs, an infrared heat gun, which relies on a material's absorption of infrared radiation rather than hot-air convection for heating, is more convenient. Infrared can produce almost

mold a shrinkable connector cover to a connector (right) without damaging the wires or connector. A clean flame or infrared can do the job in some cases.
instantaneous shrinkage, but you must be careful not to overheat the material.

So not only do you have a wide choice of shrinkable materials; you have a big choice of heating devices as well. Which one is best, of course, depends on the application, the job size, the environment, the speed, the convenience and the uniformity of results.
A number of government and commercial standards can be used to determine your choice of shrinkable material. The most widely used spec is MIL-I-23053. Although intended for military procurement agencies, the spec also has been adopted by many industrialequipment makers, so that many drawings for commercial-grade equipment call out shrinkable materials by referring to this spec.
MIL-I-23053 includes an extensive list of the shrinkable materials you may use, with dimensional data and tolerances for each material. Tests for tensile strength and resistance to solvents, oils and fungus growth are provided. And the properties of the material before and after heat aging are specified.
Some shrinkable material is recognized by Underwriters' Laboratories (subject 224). But because UL -in this case-is more conservative than the military, UL doesn't list as many types. The flammability requirements are especially severe, since UL emphasizes fire safety. To ensure traceability, UL requires that the material be surface-marked with the manufacturer's name or UL-file identification number, and a temperature rating.
But UL marking should be requested by the user. Most users don't require the UL marking, although they may use UL-type tubing to be sure of getting high performance...


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## NDRO logic circuits use Josephson effect

Josephson junctions are potential memory elements for ultra-high-speed logic circuits. But retaining information on readout has complicated the logic circuitry. Now, a nondestructivereadout (NDRO) memory cell based on the Josephson effect has been designed at the University of Karlsruhe, West Germany.

The equivalent circuit of an NDRO memory cell (Fig. 1) consists of two Josephson junctions, having unequal maximum currents, connected by an inductance. The inductance and the two junctions form an interferometer loop. Information is stored according to the presence or absence of a single flux quantum.

As with all superconducting switches, operation is fast, with pulses of the order of 100 ps . A major advantage of the NDRO cell, however, is that external bias currents are not required to preserve the stored information. In addition, the control pulses for all access lines have the same polarity and amplitude, so simple drive circuits can be used.

Individual memory cells are selected by a word line, which supplies gate

current $\mathrm{I}_{\mathrm{G}}$, and by two control lines that are inductively coupled to the cell. One control line, $\mathrm{I}_{\mathrm{cx}}$, is at right angles to the word line, $\mathrm{I}_{\mathrm{G}}$, while the other control line, $\mathrm{I}_{\mathrm{CY}}$, runs in a staircase pattern through a matrix of cells.

A read operation is performed by applying control currents $I_{G}$ and $I_{C X}$. If $I_{C X}$ is switched off before $I_{G}$ the information in the cell isn't changed by the read operation. If all three currents are applied simultaneously, a ZERO is written in the cell. To unite a ONE, only currents $\mathrm{I}_{\mathrm{CX}}$ and $\mathrm{I}_{\mathrm{CY}}$ are applied.
The cell can be easily fabricated since the inductance connecting the two junctions consists of a superconducting strip deposited on top of a thick oxide layer.

## N -way power splitter cuts insertion losses

Normally, a radio-frequency power splitter, which divides power applied to its input port equally among a number of output ports, is put together with simple two-way splitters as building blocks. But if the number of output ports is not a power of two, unused ports are terminated internally-splitter-insertion loss goes up. Now, with a technique developed at the European Space Research and Technology Center in Noordwijk, Holland, low-loss splitters with any number of output ports can be designed easily.

An N-way splitter (one input, N outputs) consists of resistors and ideal inverting transformers-with a $1:-1$

turns ratio (see Fig. 1). Each output port is connected to every other output port by a resistor of $\mathrm{NR}_{0}$, so that the number of resistors required is $1 / 2 \mathrm{~N}(\mathrm{~N}-1)$. $\mathrm{N}-1$ transformer windings are connected in series between each output port and the input port, so that $1 / 2 \mathrm{~N}(\mathrm{~N}-1)$ transformers are required.

Connections for the resistor and transformer when $\mathrm{N}-4$ are shown in Fig. 1 and Fig. 2, respectively. The complete four-way splitter is obtained by combining the two sets of connections. The input impedance is $R_{0} / N$ and each output-port impedance if $R_{0}$.

A four-way splitter has performed well over a 10:1 frequency range.

## Improved fiber-breaking 'sparked' by Post Office

Limitations of the latest technique for breaking optical fibers cleanly for splicing-spark erosion-have been overcome by modifying the spark gap and its electrodes.
When the end of an optical fiber is joined to another fiber or optical component, the fiber end must be of very high quality. But the spark-erosion technique has not always given repeatable high-quality results.
Investigations by the British Post Office have led to a more reliable spark-gap technique that allows simple mechanization of the fiber-breaking process. With the spark technique the fiber is placed between two electrodes and subjected to a high-voltage spark that is fired at the rate of a few thousand pulses per second. The fiber can be broken by hand, but for better results, it is pulled apart mechanically.
The British Post Office also found that the commonly used platinum or tungsten-carbide electrodes were inferior to copper or steel electrodes. The former wear away quickly and it is thought that the fiber is weakened by surface damage caused by accelerated metal ions.

It was also learned that the best breaks can be obtained with a diffuse spark, which is achieved with a bundle of wires, instead of with a solid electrode, on one side of the spark gap.

These two-piece, Hi-Density connectors offer the best combination of cost and performance available. They are ideal for use on single sided, double sided and multilayer boards and are well suited for applications where shock and vibration are a factor (tests show better than 200 Hz at 20 g .).
$\square$ Easy contact removeability to replace worn or broken pins
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COMECTORS


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Essex/Stancor, 3501 W. Addison St.,
Chicago, IL 60618, 312/463-7400.


## Optocoupled line-receiver input discriminates against narrow noise pulses

The simple optocoupled interface in Fig. 1a connects a data line to the receiving end of a data link that features pulse-length discrimination to enhance noisepulse rejection. A rugged red LED, $\mathrm{D}_{1}$, can bypass any reasonable fault currents to protect the relatively fragile optocoupler input diode.

Normal $20-\mathrm{mA}$ signal pulses about 1 ms wide pass about 15 mA through the LED and cause it to emit a visible flash for each pulse-a useful test feature. The LED's peak current rating-1A for $1 \mu \mathrm{~s}$-enables the circuit to handle most transients safely.

During an input pulse, capacitor C discharges with time constant $\mathrm{CR}_{4}$; during quiescent periods C recharges with a shorter time constant, $\mathrm{CR}_{3}$. If the input pulse is so short that C can't discharge below the set threshold-voltage of the circuit's set/reset bistable, the circuit remains in its set state and no output is generated. If the pulse is longer, at the end
of the input pulse the bistable is briefly reset, before being set again when $C$ recharges.
Thus, each input pulse greater than the threshold length-about $500 \mu \mathrm{~s}$ with the component values shown-generates a $6-\mu \mathrm{s}$ negative-going output pulse. This pulse can be lengthened by putting a resistortypically $47 \mathrm{k} \Omega$ for a $30-\mu \mathrm{s}$ pulse-in series with $\mathrm{D}_{2}$.

The circuit's threshold length has been found to vary from 325 to $600 \mu$ s, depending on the voltage threshold of the set input, but this variation isn't important in our application. Of course, the threshold length can be adjusted by changing the time constant $\mathrm{CR}_{4}$.

T.M. Napier, MSc., Electronics Engineer, Radiation Protection Group, CERN European Organization for Nuclear Research, 1211 Geneva 23, Suitzerland.<br>Circle No. 311



## HERFS YOUR CHANCE 10 IRY His=ROPICS.

Here for the first time is a reasonably priced, off-the-shelf fiberoptic engineering kit with all the electronic and mechanical components necessary for use in TTL systems up to 5 mbps.

Augat developed it to give engineers a quick and easy way of evaluating the exciting new technique of fiberoptic interconnection in their existing or prototype systems.
The price
give you all you need to know to use it . even assuming no prior experience in fiberoptics.

The kit contains a 5-meter length of Hytrelt-jacketed cable terminated with ferrules that

5 mbps over a temperature range of 0 to $55^{\circ} \mathrm{C}$ without drifts or inadvertent comparator switching usually associated with non-temperature referenced pre-amps.
have precision ground and polished ends. All connector
is right.* And the kits are in stock at Augat's nearly 200 worldwide distributor locations.

The combination of the kit's driver, emitter, cable assembly, pre-amp, and detector provide the necessary elements for a complete TTL-compatible digital fiberoptic system. We've even included mounting brackets and sockets for convenience. And its comprehensive instruction manual will driver and pre-amp (No 698-OK-001) S99 50
elements feature gold-plated brass construction to ensure the integrity of shielded enclosures.

The temperature referenced pre-amp operates from dc to

tDupont trademark

All components of the kit are available separately. Standard accessories include butt splices. o-ring seals, and cables of other lengths. For more details and a list of Augat distributors, write Augat, Inc., 33 Perry Avenue, P.O. Box 779, Attleboro, Mass. 02703.

## AUGAT

Augat interconnection products, Isotronics microcircuit packaging, and Alco subminiature switches.

## Ideas for design

## Upper and lower thresholds can be set independently in latching-comparator circuit

Here's a simple way to provide hysteresis in a comparator circuit that overcomes the problems of conventional hysteresis circuits. It uses two comparators and a set-reset latch (Fig. 1a). The circuit is used to "clean up" noisy inputs to digital systems.

The usual way of introducing hysteresis employs positive feedback around the comparator. But the amount of hysteresis is a function of the feedback ratio, which is affected both by input impedance and output voltage. These interactions make independent control of threshold and hysteresis difficult.

In the set-reset latch circuit, the two thresholds can be adjusted independently to the limits of the required dead band. When an input exceeds the upper threshold, the latch sets to provide a ONE output. Only an input below the lower threshold can then reset the output to ZERO. Inputs within the dead band produce low levels at both comparator outputs, and the latch isn't affected.
A fast, single-chip implementation of the latch

circuit uses a Signetics NE 521 dual comparator (Fig. 1b). The chip includes Schottky TTL-gated outputs, which are connected to form a latch.

Another advantage of the latched-comparator circuit is that it can be modified so that data are clocked into the latch only during specific intervals. Instead of a set-reset flip-flop, use a j-k flip-flop with the dual comparator (Fig. 2). Not only are data clocked, thus ignoring any noise spikes or false levels between clock transitions, but hysteresis applied to the valid input data can't be affected by such spikes. The output of the flip-flop can change only when both the input is above (or below) the dead band and a clock edge goes low.
R. S. Viles, Technical Specialist, Xerox Research (UK) Ltd., 99 Bridge Rd. E., Welwyn Garden City, Hertfordshire AL7 1LQ, England.

Circle No. 312


[^6]

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For complete facts, write Dow Corning Corporation, Dept. A-7540, Midland, Michigan 48640. Tell us about your application and we'll send a free sample.

DOW CORN/NG

## Ideas for design

## Use double-buffer characteristic of a UART and still avoid overrun errors

You can use the double-buffer characteristics of a UART to speed data communications with a simple handshake structure. When the UART in the figure receives and transfers a character to its holding register, its output labeled DAV goes high.
Output DAV remains in this state as long as a character remains in its register. However, if RDAV doesn't go low after a second character has been received, the UART may receive a third character and produce an overrun error.
The handshake circuit in the figure depends on a received acknowledge signal (ACK) to prevent an overrun. On initial power-up, $\mathrm{R}_{1}$ and $\mathrm{C}_{1}$ reset both $\mathrm{IC}_{1}$ and $\mathrm{IC}_{2}$. The first character received makes $\overline{\mathrm{RDY}} \mathrm{go}$ low, which sets $\mathrm{IC}_{2}$. A low $\mathrm{IC}_{2}$ output from $\overline{\mathrm{Q}}$ stops DAV from generating further RDY transitions until the handshake ACK line goes high and resets $\mathrm{IC}_{2}$.
The RDY signal supplements the function of the UART's RDAV signal. On receiving ACK, if there should be a character in the UART's holding register, $\overline{R D Y}$ immediately goes low again (DAV is high). Otherwise, the NAND gate $G_{1}$ is enabled and ready for the next DAV to generate a $\overline{\mathrm{RDY}}$ signal.

Ban Bong, Senior Technical Staff Engineer, Advanced Technology and Engineering Dept., Collins Government Telecommunications Group, Rockwell International, Cedar Rapids, IA 52406.

Circle No. 313


A UART in a simple handshake configuration can use the UART's double-buffer capability to speed data-handling throughput.

## IFD Winner of August 16, 1977

Mike Yakymyshyn, Edmonton Telephones, 10405-104 Ave., Edmonton, Alberta, Canada T5J-OK7. His idea "Control the Speed and Phase of a dc Motor by Comparison Against a Control Frequency" has been voted the Most Valuable of Issue Award.

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## We've just terminated your flexcircuit connector cost problems... without sacrificing reliability.

## Burndy Flexlok connectors combine high-reliability with low-cost design to slash installed costs $66 \%$.

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What's more, these savings are all yours without sacrificing reliability. That's because Flexlok connectors feature Burndy's patented GTH ${ }^{\mathrm{mm}}$ contact design that delivers gas-tight, high-pressure, good-as-gold contact even under adverse environment. Hard to believe? The proof is in the cost comparisons and performance data shown below.

| Design Simplicity | 1 piece | 2 pieces <br> or more | 2 pieces <br> or more | 2 pieces <br> or more | 2 pieces <br> or more |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Conductor Types <br> Accommodated | Round <br> Flal <br> Flex. P.C. | Flat <br> Flex. P.C | Round | Round <br> Flat | Round <br> Flat <br> Flex PC |
| Top or Side Entry <br> Available | Yes | No | No | No | No |

FLEXLOK PERFORMANCE DATA

| Conlact Resistance Tesı Data |  | $\begin{aligned} & \text { MILLIOHMS } \\ & \text { MIN MAX. } \end{aligned}$ |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | Initial Contact resistance | 7.00 | 7.60 | 7.26 |
|  | After thermal shock | 7.10 | 750 | 7.25 |
|  | After durability (5 cycles) | 7.10 | 7.80 | 7.39 |
|  | After moisture resistance (10 days) | 7.20 | 8.70 | 7.68 |
|  | After vibration | PASSED |  |  |
|  | After mechanical shock | 8.20 | 25.20 | 1230 |
|  | Insulation resistance (megohms $\times 10^{6}$ ) | 002 | 9.50 | 5.26 |
|  | Dielectric withstanding voltage No breakdown @ 500V AC | PASSED |  |  |
| ~ | Initial contact resistance | 7.00 | 7.50 | 7.25 |
|  | After thermal shock | 7.20 | 7.90 | 7.46 |
|  | Ammonium Sulfide exposure (3 min.) | 7.20 | 8.00 | 7.59 |
| $\begin{aligned} & \text { m } \\ & \text { 른 } \end{aligned}$ | Initial contact resistance | 7.10 | 7.50 | 7.25 |
|  | After gas tightness | 7.00 | 7.60 | 7.24 |

Report No G7515-755 (Summary) Mated with tin/lead plated flexible printed circuitry For details, call or write: Burndy Corporation, Norwalk, Connecticut 06856 (203-838-4444).

# Temperature controller keeps components out of the ovens 



Thermonics, Inc., 750 N. Mary Ave., Sunnyvale, CA 94086. Jim Kufis (408) 733-6122. 84950; 30 days.
Component-temperature tests come out of large temperature chambers and right to the lab or production bench with the T-2050 precision temperature forcing system. In a chamber that is just slightly larger than a component being tested, temperatures can be controlled accurately and quickly from -60 to 150 C , which is wider than the military spec range.
Tests are performed without having to modify test equipment or remove components from PC boards. Not only that, but with the T-2050, Thermonics claims to have eliminated the problems of temperature gradients, repeatability errors, moisture, frost, long leads and high-frequency ringing that occur with ovens, temperature chambers, bell jars, fluorocarbon baths and temperature probes.
Temperatures within $\pm 1 \mathrm{C}$ are achieved over the full range by directing a continuous stream of dry nitrogen over a device under test. A remote
temperature sensor close to the DUT closes the loop, and feeds the proprietary proportional-control temperature controller. So the DUT is immersed in a stream of dry gas that forces, then maintains the precise temperature set by the operator.

Test temperature is set with a knob that controls the set-point reading on the digital display. The operator sets gas flow rate, using a lockable 10 -turn pot to choose a volumetric flow rate between 100 and 600 standard $\mathrm{ft}^{3} / \mathrm{h}$, which can produce linear velocities greater than $2000 \mathrm{ft} / \mathrm{min}$.
"Because we move the gas over the DUT at a high flow rate, we can guarantee precise temperature repeatability of $\pm 0.5 \mathrm{C}$ and accuracy and stability of $\pm 1 \mathrm{C}$, says Jim Kufis, Thermonics president. "Also, the gas flow can be increased to dissipate component heat up to 60 W . Fast flow is the key to the exact repeatability of test results."

Operation is simple. Connect the $\mathrm{LN}_{2}$ container, insert the DUT in the test fixture, turn power on, and select the desired test-temperature and gas flow.

The rest is automatic.
When performing only hot tests, the system may be connected to a source of $\mathrm{N}_{2}$ or dry air instead of $\mathrm{LN}_{2}$. The digital readout on the temperature controller can be switched to show either the temperature selected or the actual DUT temperature. An analog panel meter indicates the volumetric gas flow, which determines the speed of temperature changes. The system's thermal response allows full-range temperature change, from -60 C to 150 C or vice versa, in just 7.5 min .
Designed to be compatible with virtually all manual and automatic test systems, the T-2050 can be used with several types of fixtures that incorporate the gas outlet and temperature sensor. These include a general-purpose thermal cap, a thermal test socket for ICs, and a thermal rail for component handlers, which presoaks ICs to temperature before they reach the test head.
"An unusual use is to temperaturetest hybrids before encapsulation," Kufis points out. "We may also be able to provide the ability to temperaturetest circuits while they are still in the wafer stage."
Designed for bench-top operation and movable from one work area to another, the $\mathrm{T}-2050$ weighs 40 lb and measures $17 \times 17 \times 14 \mathrm{in}$.

CIRCLE NO. 301

## Logic-state analyzer plugs into Tek scope

Scanoptik, P.O. Box 1745, Rockville, MD 20850. Jerry Shumway (301) 977-9660. \$3250; 8 wks.
Logic-state analyzer, LC-732, plugs into Tektronix 7000 oscilloscopes. The single-module unit has 32 -channel capability and fits in a four-wide scope without affecting normal scope operation. For triggering, the device matches a 16 -bit $\mu \mathrm{C}$ address bus or 8 -bit data bus, or both combined for a full 24 -bit word trigger. After address match, there is a digital delay capability up to 65,000 counts. The memory stores 64 words of 32 inputs, and presents the information in hexadecimal characters on the scope.

CIRCLE NO. 303

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## Adapter converts scope to logic analyzer

Mid-South Instrument Services, P.O. Box 1252, Gretna, LA 70053. (504) 393-0450. \$74.95; stock.

The MS-1 multiplexer switch is a compact adapter for converting any single channel oscilloscope into a multichannel logic analyzer. The adapter features 2,4 or 8 channels of displayed data which are switch-selectable, and it operates in either chop or alternatesweep modes. Up to eight data lines may be sampled, while displaying the digital-logic levels and timing relationships on the scope. Each input channel is multiplexed through an analog switching device, so that waveforms as well as logic levels are preserved. Size is $51 / 2 \times 3 \times 11 / 4 \mathrm{in}$. With a 9-V battery, the current drain is less than 6 mA .

CIRCLE NO. 304

## Digital-pulser probe boasts bit of automation



Continental Specialties, 44 Kendall St., New Haven, CT 06509. (203) 624-3103. $\$ 74.95$.

In digital pulser, DP-1, internal circuitry monitors the node being probed, then presets the dual-mirror output circuitry to pulse the node the other way. The probe delivers a $50-\mathrm{mA}$ pulse in the CMOS mode, 100 mA in the TTL mode, to kick most lines with no need to desolder, unplug or isolate. Power is derived from the circuit being tested, and a switch selects threshold levels to trigger either CMOS or TTL. If the pulse pushbutton is held down for more than a second, the probe delivers trains of about 100 pulses/s.

CIRCLE NO. 305

## Frequency counter has period mode, too


$B \& K$-Precision, 6460 W. Cortland Ave., Chicago, IL 60635. Myron Bond (312) 467-1326. \$450; stock.

Model 1850 frequency counter measures from 5 Hz to 520 MHz with a period measurement capability from 5 Hz to 1 MHz . The counter is fully autoranging in either auto or prescale modes, with automatic decimal point position and $\mathrm{MHz} / \mathrm{kHz}$ readout. The display is a six-digit, 0.43 -in.-high LED, with leading-zero blanking. A TCXO time base is standard for 1-ppm stability over a 0 to 50 C range. For period measurements, resolution is 1 ns. The function switch selects $\mu \mathrm{s}$ ( $100-$ period average) or auto-display reading. In auto, a one-period or a 10 or 100period average is selected.

CIRCLE NO. 306

Gen pulses to 10 MHz with 5-ns rise/fall time


Dytech, 2725 Lafayette St., Santa Clara, CA 95050. (408) 241-4333. \$185; stock to 4 wks.

Pulse generator, Model 750, has rise and fall times of less than 5 ns over a range of 10 Hz to 10 MHz . The output is variable from 0 to 10 V across a high impedance and 0 to 5 V across a 50 $\$ 2$ load. Pulse widths and delays are adjustable from 50 ns to 100 ms . The trigger output is a fixed-width square wave of less than 50 ns and 5 V , open circuit. Both the external and gate inputs can be driven by low-power logic, an ac signal or by another $50-\Omega$ generator.

Resolve down to $0.1 \mu \Omega$ with digital ohmmeter


Valhalla Scientific, 7707 Convoy Ct., San Diego, CA 92111. (714) 277-2732. \$2495; stock.
The Model 4275 digital ohmmeter resolves $0.1 \mu \Omega$. Measurement range covers $1 \mathrm{~m} \Omega$ to $100 \Omega$. Test current on the most sensitive range is 10 A with a minimum compliance voltage of 10 V dc. Six decade ranges are pushbutton selectable with $100 \%$ overrange capability. Heavy-duty, industrial-grade, gold-plated Kelvin clip cable sets terminated in Bendix connectors are available in lengths up to 20 ft . Options include automatic temperature compensation, BCD outputs and rack adapter.

CIRCLE NO. 308

## 16-channel digital latch mates with Tek scope



Tektronix, P.O. Box 500, Beaverton, OR 97077. Wyn Giluck (503) 644-0161. \$1475; 4 wks.
The company's logic-analyzer line using 7000 series oscilloscopes has been extended with the 16 -channel DL2 digital latch. This module provides the designer using a 7D01 logic analyzer the ability to latch pulses occurring asynchronously to systems activity. The pulses can be as narrow as 5 ns at the probe tip, with an amplitude as small as 500 mV centered on a threshold set by the user. The P6451 data-acquisition probes connect to the DL2 which then interfaces to the 7D01.

CIRCLE NO. 309

# If you think logic analyzers, recorders or scopes are the only way to debug digital circuitry, yourewrong. 

## Wellshow you. guesswork out of debugging the complex circuitry you've designed. Makeit less of an art <br> You'd like to have a known, stable signal to input

 to your circuit, so you can tell whether its output is on target. But the multiple pulse or wave generators, flip flops, logic gates or other gear you've been using to generate your word and timing patterns just aren't enough. You can't program them easily or with any guarantee they're accurate.You'd rather have a single instrument that has multiple channels and is easily programmed. So you can input exact duplicates of the real programs your circuit will be handling. So you can tell at a glance whether your output is right.

But nobody makes a single unit like that, right? Wrong!

Interface Technology is going to show you.

We'll show you a self-contained, low-cost, small digital debugging tool for $\$ 2,500$ and up. A multichannel, microprocessor-controlled data and timing generator that's programmable with only 16 instructions. A general purpose tester that can generate large amounts of digital data, that's interactive, can respond to external stimuli and make decisions. A benchtop instrument that can be used rack-mounted as part of a computer-driven system, or as the core of a low-cost, stand-alone digital test system.

We'll show you how our multi-channel programmable digital systems can lower your design time. We'll show you how you can be sure about your final product. And we'll show you how we're already working for people like General Dynamics, Rockwell and Hughes and how we can work for designers like you.

We'll put the whole show on the road and demonstrate our digital test systems for you in person. And all you have to do is call collect: (213) 966-1718. Or write to the address below.
> interface
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CIRCLE NUMBER 46

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| :---: | :---: | :---: | :---: | :---: | :---: |
| 71850 | $\begin{aligned} & 100 \mathrm{kHz} . \\ & 3 \mathrm{GHz} \end{aligned}$ | BNC <br> Male | BNC Fem | $\pm 0.5 \mathrm{~dB}$ | 70 |
| 73N50 | $\begin{aligned} & 100 \mathrm{kHz} \\ & 4 \mathrm{GHz} \end{aligned}$ | N Male | BNC Fem. | $\pm 0.2 \mathrm{~dB}$ | 75 |
| 74N50 | $\begin{aligned} & 10 \mathrm{MHz} \\ & 12.4 \mathrm{GHz} \end{aligned}$ | N Male | $\begin{aligned} & \text { BNC } \\ & \text { Fem. } \end{aligned}$ | $\pm 0.5 \mathrm{~dB}$ | 145 |
| 74S50 | $\begin{aligned} & 10 \mathrm{MHz} \\ & 12.4 \mathrm{GHz} \end{aligned}$ | SMA <br> Male | $\begin{aligned} & \text { BNC } \\ & \text { Fem. } \end{aligned}$ | $\pm 0.5 \mathrm{~dB}$ | 165 |
| 75A50 | $\begin{aligned} & 10 \mathrm{MHz} \\ & 18.5 \mathrm{GHz} \end{aligned}$ | APC. 7 | BNC Fem | $\pm 1 \mathrm{~dB}$ | 190 |
| 75N50 | $\begin{aligned} & 10 \mathrm{MHz} \\ & 18.5 \mathrm{GHz} \end{aligned}$ | N Male | BNC Fem. | $\pm 1 \mathrm{~dB}$ | 170 |
| 75S50 | $\begin{aligned} & 10 \mathrm{MHz} \\ & 18.5 \mathrm{GHz} \end{aligned}$ | SMA <br> Male | BNC Fem. | $\pm 1 \mathrm{~dB}$ | 170 |



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## DATA PROCESSING

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Tri-data, 800 Maude Ave., Mountain View, CA 94043. Melinda Magnett (415) 969-3700. \$1995.
FlexiFile, a flexible disc system, can be used for program loading, data logging, and on-line data collection, and is fully user-programmable from its front panel. A text-editing software package is included which features insert and delete, file merging and character-string search. Equipped with the RS232 and/or current loop interface, the device will accommodate a terminal and modem simultaneously at independent baud rates. The user can prepare his data off-line and then batch-process it to a CPU at a high transfer rate.

CIRCLE NO. 320

## Hi-density tape systems mate with CDC computers

Control Data, P.O. Box 0, Minneapolis, $M N$ 55440. (612) 853-7600.

A series of high-performance magnetic tape subsystems that record data at densities up to 6250 bytes $/ \mathrm{in}$. is available for CDC's large-scale computers. The CDC 677 and 679 tape transports and 7021 controllers are for use with the CYBER 170, CYBER 70, and 6000 -series computer systems running under NOS/BE and NOS operating systems. Six 9 -track models of the 679 tape transports operate at speeds of 100,150 or $200 \mathrm{in} / \mathrm{s}$. Three models are offered with 800 or $1600 \mathrm{bit} / \mathrm{in}$. densities, NRZI or phase-encoded (PE) recording. Three units feature 1600 bit/in. PE recording, or group-coded recording, at a density of 6250 bytes/in. All units provide data transfer at rates of 160,000 to 1.25 million char/s.

CIRCLE NO. 321

Tablet digitizer converts graphic data to digital


GTCO, 1055 First St., Rockville, MD 20850. (301) 279-9550. \$750 up (large qty).

The Datalyzer tablet digitizer, when used with a cursor and electronic controller, converts graphic data into a digital format for computer processing. The device provides graphic data inputting similar to a keyboard operation for inputting alphanumeric data. A crystal-controlled electromagnetic scanning method is used with a freemovement cross wire/pen cursor. Graphics can be digitized from source materials up to a thickness of 1 in . Standard tablets have an $11 \times 11,14$ $\times 14,11 \times 17$ and $20 \times 20$-in. active area.

CIRCLE NO. 322

## Any popular mini reads these diskettes



Techtran Industries, 200 Commerce Dr., Rochester, NY 14623. (716) 334-9640.

Type 9512 micro-discs recorded on the Model 9512 recorder from any RS-232 terminal or data logger can be read on any popular minicomputer disc system. Diskettes generated on-line from the DEC, RCA or Intel systems can be read off-line by the micro-disc, back to an ASCII terminal or, via the RS-232 communications interface, to a remote location. The diskettes provide total random access by character string, easy editing, bidirectional skip and economical data storage and retrieval.

CIRCLE NO. 323

## Video system is crammed onto $5 \times 10$-in. board

IOR, P.O. Box 28823, Dallas, TX 75228. (214) 358-2671. \$399; stock.

The VDS2K video system contains all address decoding necessary for both memory-mapped and isolated I/O on a $5 \times 10-\mathrm{in}$. PC board that is fully S100bus compatible. Video output is either XYZ-TTL compatible or composite video. A Greek, upper/lower case character with a $7 \times 9$ matrix in a $9 \times 12$ field is displayed with attribute availability. The actual display consists of 1920 char organized as 30 lines of 64 char plus a 128 -char, nondisplayable buffer. There is 2 k of memory with under $500-\mathrm{ns}$ read/write time. A screen of data can be changed in under 800 $\mu \mathrm{S}$.

CIRCLE NO. 324

## Cluster system accesses local databases fast

Delta Data Systems, Woodhaven Industrial Park, Cornwells Heights, PA 19020. Barry Maser (215) 639-9400.

The Model 6500 cluster-computer system is for storage and immediate access to local databases in a standalone or a distributed processing environment. The system permits up to eight video display terminals to share access to a processor, a host communications port, up to $1.5 \times 10^{6}$ char of storage and up to two printers.

CIRCLE NO. 325

## Desktop unit scrambles for data security

Datotek, 13740 Midway Rd., Dallas, TX 75240. (214) 233-1030.

A portable, self-contained, desktop encryption unit, DC-26, has a full-size keyboard and thermal page printer. As the message to be enciphered is typed, the characters appear on a five-character LED display. By continued typing, the characters are shifted off the display, enciphered and printed. Messages are deciphered by entering the scrambled characters in five-character groupings, with the characters being successively shifted off the display, deciphered and printed. A key generator provides an almost infinite number of user-selected key variables ( $10^{52}$ possibilities), making the possibility of unauthorized interception extremely remote.

CIRCLE NO. 326

# We'll show you: 

PACIFIC COAST

Ward/Davis Associates
Lawndale, CA 213-542-7740
La Jolla, CA 714-459-3351
Sunnyvale, CA 408-245-3700

## MID-ATLANTIC

Eastern Instrumentation of New York
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Eastern Instrumentation of Philadelphia
Philadelphia, PA 215-927-7777
Naco Electronics
Syracuse, NY 315-699-2651
Utica, NY 315-732-1801
Fairport, NY 716-223-4490
Creative Marketing Associates
McLean, VA 703-893-6612

## ROCKIES/SOUTHWEST

BFA Corporation of Arizona Scottsdale, AZ 602-994-5400 Denver, CO 303-837-1247 Albuquerque, NM 505-292-1212 Las Cruces, NM 505-523-0601 Salt Lake City, UT 801-466-6522

## NEW ENGLAND

Instrument Dynamics
Wakefield, MA 617-245-5100

## SOUTH CENTRAL

Data Marketing Associates
Houston, TX 713-780-2511
Dallas, TX 214-661-0300
San Antonio, TX 512-828-0937
Austin, TX 512-451-5174
Norman, OK 405-364-8320

## MIDWEST

Carter Electronics, Inc. Minneapolis, MN 612-559-1976 Chicago, IL 312-585-5485 Indianapolis, IN 317-293-0696 Milwaukee, WI 414-464-5555 St. Louis, MO 314-569-1406 Overland Park, KS 913-649-6996

[^7]
# Bit-slice development system looks like a compact instrument 



Step Engineering, 154 San Lazaro Ave., P.O. Box 61166, Sunnyvale, CA 94086. Steve Drucker (408) 733-7837. P\&A: see text.

An instrument for developing bitslice microcomputers combines the functions of several ROM simulators, a logic analyzer, and a command-language CRT terminal in a single 5-1/4 $\times 18 \times 20-\mathrm{in}$. package. The STEP-2 Microcode Array Simulator and Processor Diagnostic System is a new type of tool for checking out real-time firmware/hardware, diagnosing faults, and patching microcode.

Step Engineering prefers to call the STEP-2 an instrument, not a development system, because it has a fixed program, it's much more compact than earlier units, and it's easier to operate. Learning its 14 English-language interactive commands is just as easy as learning the functions and the controls of a scope or logic analyzer, according to the company.

The STEP-2 is small because it performs no software tasks. Instead it
concentrates its power on firmware/hardware checkout. Moreover, where doing an assembly would tie up the typical development system and halt hardware tests, the STEP-2's hardware checkout won't be disrupted because the software is developed elsewhere, in a mini or large computer, and transmitted to the STEP-2 through the RS232 interface. Also, a new assembly can be loaded down-line from another computer when it's needed.
To simulate the macrocode and microcode memories of the application hardware, the memory can be partitioned into two independent arrays that operate simultaneously. Up to 96 kbits of user memory in STEP-2 can be switch-configured efficiently for widths of 8 to 96 bits and depths of 1 to 12 k , and cycled at 20 MHz .

Plug-in memory interface cards provide three options to optimize connection to the user's system: nonbuffered for high-speed direct connection, buffered and latching. A wrapped-wire
area on the interface card permits changes in bit mapping to ease the transition from wrapped-wire prototype to PC layout since the pin assignments usually change.
Any bipolar ROM can be emulated by the STEP-2 regardless of ROM organization. Optional ROM emulator plug-ins enable the STEP-2 to interface directly with existing user PROM sockets. The STEP-2 user interface consists of a $5-\mathrm{in}$. CRT display and alphanumeric keyboards. The display features an 8 -line $\times 32$-character fully formatted, word-oriented output with 1/4-in. characters.

A full ASCII keyboard with separate cursor-control keypad and hex dataentry keypad controls the instrument, manipulates data and communicates with external computers.

Input to the STEP-2 comes through a serial-communications link, RS232, or a $20-\mathrm{mA}$ current-loop, at 110 to 9600 baud. And with a modem interface, the STEP-2 can be hooked up to a timesharing facility over a phone line.

A word-oriented object-code editor (in ROM) permits address information and object code to be displayed in either octal or hex. An "expand" feature permits a selected portion of object-code word to be displayed and modified in binary.

Several diagnostic outputs are provided to help with system debugging. What's more, processor activity can be monitored directly on the CRT. Multiple real-time breakpoints allow conditional-jump analysis. The breakpoint state is displayed on the CRT and duplicated on BNC connector outputs.

Program execution can be monitored with a real-time trace, which includes combinatorial breakpoint triggers with a cycle counter and programmable relationship between trace length and trigger.

Delivery is scheduled to begin this month with the real-time trace option to be available the first quarter of 1978. Prices start at $\$ 9000$, which includes 64 kbits of high-speed reconfigurable memory, and run to $\$ 14,000$, with all options.

CIRCLE NO. 327

## Add-on memory provides 2 Mbytes for PDP-11/70

Monolithic Systems, 14 Inverness Dr. E., Englewood, CO 80110. Karen Garritano (303) 770-7400. \$9525; 6 wk.
An add-on memory, MSC 3602, for the DEC PDP-11/70, provides up to 2 Mbytes in a self-contained unit. It is expandable in 64 -kbyte increments to 4 Mbytes with an additional $51 / 4-\mathrm{in}$. chassis. Access time of 500 ns and cycle time of 700 ns permit using the PDP-11/70's memory bus at the maximum data rate. MSC 3602 is a $10^{1 / 2-}$ in. high freestanding or rackmount chassis that includes power supply and forced-air cooling.

CIRCLE NO. 328

## I/O card provides real-time clock

Hewlett-Packard, 1501 Page Mill Rd., Palo Alto, CA 94303. Mike Malone (415) 493-1501. \$600; 4 wks.

An I/O card provides both a realtime clock and timing generator for use with the HP 9825A desktop computer. The HP 98035A card provides real-time information in months, days, hours, minutes and seconds. The device can interrupt at a specific time, after a time delay or a periodic interval. A counter is incremented every millisecond and an internal battery holds real time for up to two months. Error checking and external triggering is provided.

CIRCLE NO. 329

## Controller has remote switch for 16 channels

International Data Sciences, 100 Nashua St., Providence, RI 02904. (401) 274-5100. \$1090; $4 w k$.
Model 8962 controller and power module provides individual remote control for 16 channels of $A, B$ switching when used with the Model 8931 remote control panel. It drives eight Model 8909 or eight Model 8914 two-channel modules used for frontend processor A, B switching of EIA or telephone line interfaces. The master control switches all 16 channels to A or B. Individual A, B switching is provided on each switching module or at the 8931 panel. LEDs display the status of 12 key signals.

CIRCLE NO. 330

## $\mu \mathrm{C}$ card can be easily expanded



CGRS Microtech, P.O. Box 368, Southampton, PA 18966. (215) 757-0284.

With the CGRS "Microputer 6000," S-100 bus users can take advantage of a standard packaging scheme and have the high speed and versatility of the 6502 . The computer card contains 4 k of EPROM (2708) and 2 k of RAM (2111) in addition to the 6502 microprocessor and TTL support logic. The unit can be easily expanded with an available line of support products including a front panel, I/O cards, packaging and software.

CIRCLE NO. 331

## GPIB CONTROLLER and Matrix Switching System

 designed specifically for IEEE BUS applications! Controls any IEEE-488 compatible instrument! Universally accepted high-level language BASIC! Microcomputer with up to 56 K byte memory! RS232C/TTY current loop interface standard!


For more information, call, write or circle the reader service number below. Systron-Donner Data Products Division, 935 Detroit Avenue, Concord, CA 94518. Phone (415) 798-9900. Abroad, contact Systron-Donner in Munich, Leamington Spa, U.K., Le Port Marly, France, and Melbourne.
-SYSTRON

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CIRCLE NO. 142

## CRYSTAL CLEAR EPOXY CASTING RESIN



Several transparent Stycast © resins are offered for making display embedments or castings. A convenient chart is available to aid in selection of the most appropriate system. It is yours for the asking.

CIRCLE NO. 143

[^8]
## MICRO/MINI COMPUTING

## Interface lets PDP-8 act like PDP-11



Digitek, 5950 Sixth Ave., Seattle, WA 98108. Frank Mauger (206) 762-3933. \$960; 4 to 8 wks.
A logic card, Model DK 8/11, provides the PDP-8 with a 16 -bit parallel interface. The card allows direct compatibility between the DEC PDP-80mnibus and peripherals and systems that normally would interface to the DEC PDP-11 DR11C general purpose interface. The device converts the PDP-8 12 -bit words into 16 -bit DR11C-compatible words. It plugs into one Om nibus slot and is completely compatible with the DEC DR11C interface, including all necessary control and handshaking signals so that a PDP-8 can communicate directly with a 16 -bit minicomputer.

CIRCLE NO. 332

## S-100 bus board gives peace and quiet

Thinker Toys, 1201 Tenth St., Berkeley, CA 94710. George Morrow (415) 527-7548.

An S-100 bus board in kit form, Wunderbuss, produces signals that are "textbook clean." The high signal quality of the $\mathrm{S}-100$ bus is produced by a complex noise-control system called Noiseguard. Signal isolation is achieved by a cross-coupled system of ground lines that are interlaced between signal lines, and cross-coupled ground planes. This system surrounds each signal in a cocoon of extremely quiet space to eliminate noise and cross-talk between signal lines. Each data line is actively terminated by a circuit that absorbs signal reflections and noise.

## Nonvolatile CMOS RAM allows memory expansion

Wintek, 902 N. 9th St., Lafayette, IN 47904. (317) 742-6802. \$339.

Microprocessor memory expansion with nonvolatile memory is possible with the CMOS RAM/Battery module. Memory is retained during power-off conditions including when the module is unplugged from the system bus. Two size AA nickel-cadmium batteries allow for power-off periods of up to one year. The module can accommodate up to 2 kbytes in multiples of 256 bytes and has write protection. The module comes on a $4.5 \times 6.5$-in. 44 -pin PC board.

CIRCLE NO. 334

## Static RAMs mate with S-100 bus

Dynabite, 4020 Fabian Way, Palo Alto, CA 94303. Mike Watts (415) 494-7817. $\$ 525$ to $\$ 995$; stock.

Modules of $16-\mathrm{k}$ and $32-\mathrm{k}$ static RAMs are available with access times of 450 or 250 ns . The $250-\mathrm{ns}$ units are compatible with $4-\mathrm{MHz}$ Z-80 processors. Both $16-\mathrm{k}$ RAM modules feature bank select, allowing up to eight separate banks of up to 64 k to reside in the same system. The module may be addressed in four separate 4-k blocks along $4-\mathrm{k}$ boundaries. Each of these blocks may be individually write protected. The $32-\mathrm{k}$ modules offer $4-\mathrm{k}$ boundary addressing and complete buffering.

CIRCLE NO. 335

## Add-on parity memory mates with PDP-11's

Ampex, 200 N. Nash St., El Segundo, CA 90245. Clyde Cornwell (213) 640-0150. $\$ 4950$ to $\$ 12,750 ; 4$ wks.

An add-on parity memory, ARM-1100P, provides 32 to 128 kwords of memory for any DEC PDP-11 computer using a Unibus structure. The device can expand memory from any 8 -k boundary in 32 -k increments up to the maximum of 128 kwords . The memory uses only one CPU slot and needs just one Unibus load. Each unit is supplied with wired card rack, power supply, interface, parity control, cooling fans and interconnecting cables. An access time of 375 ns and a full cycle time of 700 ns is provided.

CIRCLE NO. 336

## POWER SOURCES

## Supplies deliver 1 kW with high efficiency



Kepco, 131-38 Sanford Ave., Flushing, NY 11352. Paul Birman (212) 461-7000. \$590; stock.

Ferroresonant power supplies, Type PRR, provide efficiencies of 70 to $80 \%$ with $1-\mathrm{kW}$ output at 12 to 48 V dc . The units have automatic current limiting, brownout protection, built-in blower and recessed metering. Line regulation at $120-\mathrm{V}$ input is $\pm 1 \%$ max for $\pm 15$ V line change at full load. Output voltage increases 0.75 to $1.5-\mathrm{V}$ for 100 to $50 \%$ load change depending on output voltage rating. A $+1 \%$ change in line frequency produces about $+1.5 \%$ output-voltage change.

CIRCLE NO. 337

## HV supply regulates current and voltage



Advanced High Voltage, 14532 Arminta Ave., Van Nuys, CA 91402. John Richardson (213) 997-7222. From $\$ 680$; $s$ to $4 w k$.

High-voltage dc-to-dc converters with outputs to 18 kV at 80 W (Series 1300) are voltage and current regulated. Remote programming of voltage and current is a feature as is the selfcontained adjustment of voltage and current. Other features include line and load regulation of $0.1 \%$ and an adjustable switching frequency to minimize specific interference. Input voltage is 24 to 32 V dc . Size: $5 \times 4.75 \mathrm{in}$. Weight: 5 lb .

CIRCLE NO. 338

# Memory Power Eternacell ${ }^{\circledR} 10$ year lithium primary battery for semiconductor memories 

Don't risk memory failure. Eternacell® high reliability, lithium primary batteries are the ideal standby power source for all types of volatile memory applications. The reasons:

- Steady voltage ( 2.9 volts per cell) at low continuous current
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Or call (914) 699-7333



## Four wàvs to build better Power Chokes... with ferrite cores from Fair-Rite

When designing switch mode power supplies, you can provide more stable inductance for the filters in the presence of DC load currents, by using these types of ferrite cores:


Gapped E-cores, giving moderate shielding at lower cost.

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4 .
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Square bobbins; most economical; without shielding properties, but least affected by DC

Write for FREE application note on use of the Hanna curve to select the proper air gap for ferrite cores. All, and much more, from

[^9]
## ICs \& SEMICONDUCTORS

## Regulator delivers 3 to 30 V and limits currents to 1.8 A

SGS-ATES, 79 Massasoit St., Waltham, MA 02154. Ruben Sonnino (617) 891-3710. P\&A: See text.

An adjustable out-put-voltage range of 3 to 30 V on a regulator is nice enough, but the L200 also offers an adjustable current limit of 0 to 1.8 A. Housed in the new five-lead TO-220 package, the monolithic regulator can be mounted directly to a heat sink for good thermal conductivity. Input voltages can range from about 5 to 40 V .

Regulation is within $0.1 \%$ of $\mathrm{V}_{\text {out }}$ for load changes up to 1.5 A , and $0.1 \%$ for input-voltage changes of 10 V . Ripple rejection is greater than 70 dB for a $10-$ $V$ change in $V_{\text {in }}$ and generated noise is typically less than $40 \mu \mathrm{~V}$ over 10 Hz to 100 kHz . An on-chip band-gap element not only provides a low reference voltage, but reduces noise as well.
During load variations, thermal drift of the output current is 100 $\mathrm{ppm} /{ }^{\circ} \mathrm{C}$, and current stability is better than $0.2 \%$ of $\mathrm{I}_{\text {out }} / \mathrm{V}$. During no-load operations, the regulator draws just 5 mA of quiescent current (ata $20-\mathrm{V}, \mathrm{V}_{\text {in }}$ ).


Thermal resistance from junction to ambient is $50 \mathrm{C} / \mathrm{W}$ and from junction to case, $3 \mathrm{C} / \mathrm{W}$.

In 1 to 99 quantities, the L200 costs $\$ 4.20$. Delivery takes from 4 to 6 weeks.

CIRCLE NO. 302

Molded triacs claim
highest power rating


Intermational Rectifier, 233 Kansas St., El Segundo, CA 90245. (213) 322-3331. $\$ 1.5 .48$ to $\$ 80.28$ ( 100 qty); 4 to 6 wks.

Molded power triacs for operation to $1200 \mathrm{~V}, 50$ and 100 A rms , with surge ratings to 900 A , are said to offer the highest power ratings of any available. The molded triac assemblies have an isolated metal base plate, isolated to 2000 V rms, that improves heat transfer. Type T50AC units have current ratings of 50 A rms with a $400-$ A surge rating; T100AC units are for 100 A rms, with $900-$ A surge rating. Both types are available with ratings of $400,600,800,1000$ or 1200 V. Junction operating temperature range is -25 to +125 C for all units.

CIRCLE NO. 339

SHERLOCK OHMS \& DR. WATTS


Timer/sequencer delays from $\mu \mathrm{s}$ to months


Exar Integrated Systems, 750 Palomar Ave., Sunnyvale, CA 94086. Brooks Hamilton (408) 732-7970. \$0.90 (100 $q t y)$.
A long-range timer/sequencer IC capable of time delays from microseconds to months (designated the XR-2242) contains a precision time base oscillator and an 8-bit binary counter. Upon triggering, it provides an accurate time delay of 128 times an RC constant. Since it uses the countdown principle, cascading several timer circuits causes the total time interval to increase geometrically. The time base oscillator is accurate to $0.5 \%$ and has a tempco of less than 100 ppm $/{ }^{\circ} \mathrm{C}$.

CIRCLE NO. 340

## Clock signals provided at up to 15 rates

Harris Semiconductor, P.O. Box 883, Melbourne, FL 32901. (305) 724-7430. See text; stock
CMOS programmable bit-rate generators, HD-4702 and HD-6405, provide clock signals for digital data-transmission systems. The HD-4702 can be programmed for one of 13 bit rates while the HD-6405 extends to 15 selectable rates. The HD-4702 has TTLcompatible pull-up circuitry and is pinout and specification identical to 34702 devices. The HD-6405 has no pull-up circuitry but offers standard high-impedance CMOS inputs. The HD-4702 and 6405 operate at 2.4576 MHz and dissipate only 4.5 and 4 mW , respectively. The devices come in 16 -pin plastic or ceramic DIPs in either -40 to $+85 \mathrm{C},-55$ to +125 C ranges. 100 quantity prices are $\$ 9.40$ for the plastic; $\$ 15.16$ for the ceramic and $\$ 20.90$ for the ceramic hi-temp.

CIRCLE NO. 341

## Receiver IC meets IBM 360/370 I/O spec

Texas Instruments, P.O. Box 5012, M/S 308 (attn: SN75128/129), Dallas, TX 75222. Dale Pippenger (214) 238-2165. $\$ 2.33 / \$ 3.10$ (100 qty); stock.
Two new eight-channel line receiver ICs meet the IBM $360 / 370$ I/O specification. Type numbers SN75128 and SN75129 feature common strobes for each group of four receivers. The

SN75128 has an active-high strobe; the SN75129, an active-low strobe. Schot-tky-diode-clamped transistors allow low current requirements, while maintaining fast typical switching speeds of 16 ns . The receiver input resistance is specified from 7 k to 20 k and input threshold from 0.7 V to 1.7 V . The ICs are output compatible with DTL and TTL. The devices are offered in either 20-pin plastic or ceramic DIPs.

CIRCLE NO. 342


Bullt-In Reliabillty
The AT-4641 features a superior gold metallization system and hermetic packaging. All Avantek transistors are manufactured under the most stringent quality controls, assuring the designer of high MTBF's.
Fast Delivery
Concerned about a supplier located halfway around the globe?
Avantek is just a phone call away. Orders placed by noon will normally ship the same day.



## Dependable Performance

The AT-4641 is a proven performer in critical military and space applications with HI REL screening available as a standard option. It features a very high dynamic range among its impressive specifications. Reliability, performance, and immediate availability - Avantek's AT-4641 is your solid choice.

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## ICs \& SEMICONDUCTORS

## One-chip data system can save a bundle



National Semiconductor, 2900 Semiconductor Dr., Santa Clara, CA 95051. Jerry Zis (408) 737-5831. \$19.95 (100 qty); stock.

A one-chip data-acquisition system can replace $\$ 100$ to $\$ 200$ worth of hybrid and discrete-component analog boards. Included on a single $28,000-\mathrm{sq}$ mil CMOS chip (ADC0816) are a true 8 -bit a/d converter with latched outputs, a 16 -channel expandable multiplexer with address input latches, provision for handling external signal conditioning and all the logic control needed for mating the chip to all standard $\mu$ Cs. The 40 -pin device performs a conversion in $50 \mu$ s. At 25 C , the linearity, zero error and full-scale error are less than $\pm 0.5$ LSB each.

CIRCLE NO. 343

## TTL-to-MOS drivers have three-state outputs

Texas Instruments, P.O. Box 5012, M/S 308 (Attn; SN75357), Dallas, TX 75222. Dale Pippenger (214) 238-2165. \$1.41 to $\$ 2.47$ (100 qty); stock.

Two quadruple TTL-to-MOS driver ICs, SN75357 and SN75375, feature three-state outputs. The 75375 has individual supply voltages for each of the four drivers, capable of operating from 5 to 24 V . The 75357 has low transient current during switching, making it useful for mating with CMOS systems. It has a high-level output voltage variable from 5 to 13.2 V . Individual supply voltage pins on the 75375 allow individual adjustment of $\mathrm{V}_{\mathrm{OH}}$ levels to match various load conditions. Typical propagation delay is 31 ns and the outputcurrent drive capability is 150 mA . The circuit has two NAND and two inverting drivers. Both circuits operate from 0 to 70 C and are in plastic ( N suffix) or ceramic (J suffix) DIPs.

CIRCLE NO. 344

Gold helps produce rugged vhf power devices


Motorola, P.O. Box 20912, Phoenix, AZ 85036. Alan Wagstaffe (602) 244-6394. $\$ 11$ to $\$ 39.50$; stock.

Dissimilar metal interfaces, which can impair reliability of military grade rf power transistors, are eliminated in the MRF314-317 series of vhf devices. Gold chip metallization, gold wirebonds and gold-plated package interfaces produce ruggedness suitable for new wideband, multimode vhf systems. The $28 \mathrm{~V}, 30$ to $100-\mathrm{W}$ series offers gains from 9 to 10 dB at 150 MHz , and is characterized from 30 to 200 MHz . Ruggedness is assured by $100 \%$ testing to withstand a load VSWR of 30:1 at rated output power.

CIRCLE NO. 345

## Single-chip $\mu \mathbf{C}$ aims at high-volume control use



Intel, 306.5 Bowers Ave., Santa Clara, CA 95051. Rob Walker (408) 249-8027. $\$ 3$ (OEM qty).

The 8021 single-chip $\mu \mathrm{C}$ is a low-cost general-purpose IC for high-volume control uses. The chip contains an 8bit control processor, 64 bytes of readwrite data memory, 1024 bytes of program memory, $21 \mathrm{I} / 0$ lines and all other generally required functions, including a programmable interval timer and event counter, system clock and oscillator. Operation is from a 4.5 to $6.5-\mathrm{V}$ supply.

CIRCLE NO. 346

## Logic arrays are field programmable

Texas Instruments, P.O. Box 5012, M/S 308 (Attn: S330), Dallas, TX 75222. Gerald McGee (713) 494-5115. \$9 (100 qty); stock.
Schottky TTL field-programmable logic arrays (FPLA), the SN54S/74S330 and SN54S/74S331, feature a built-in capability of multidimensional expansion of their basic 12 input $\times 50$ product term $\times 6$ output organization. A special circuit is included that can decode true product terms to automatically enable the FPLA outputs. The result is expandability to virtually any size array without external logic or control. This control option is activated by fusing a single link. Additionally, the S330/S331 can be programmed to stand alone; i.e., the outputs are constantly enabled when system power is applied. The arrays come in $20-$ pin DIPs.

CIRCLE NO. 356

## Power driver operates at 80 V and 300 mA peak



Dionics, 65 Rushmore St., Westbury, NY 11590. Manny Sussman (516) 997-7474. $\$ 1.38$ ( 10,000 qty); stock.

DI-446 is a universal dual high-voltage, high-current power driver in a 16 pin plastic DIP operating at voltages to 80 V and peak currents to 300 mA . The circuitry allows the device to operate with either "positive-true" or "negative-true" inputs from logic systems. The logic threshold voltage can be adjusted, thereby providing excellent noise immunity for input signals having an absolute voltage level between $\pm 40 \mathrm{~V}$. A fully isolated highcurrent diode provides transient suppression if an inductive load is to be driven.

CIRCLE NO. 357

## Zener diodes made to user specs



American Power Devices, 7 Andover St., Andover, MA 01810. Bob Dimodana (617) 475-4074. $\$ 0.065$ to $\$ 0.30$ (1000 qty).
A series of 2.5 to $5-\mathrm{W}$ zener diodes can be made to customer specification. Specifications with tolerances as close as $\pm 1 \%$ and zener voltages ranging from 3.3 to 62 V can be met. The diodes have silicon-oxide-passivated junctions, and are produced in epoxy and double plug packages. Planar oxide processing results in diodes with low leakage current, uniform sharp reverse breakdown voltages, and uniform heat dissipation. DO-7, DO-35, AIEE and AIRX axial lead and glass transfermolded packages are available.

CIRCLE NO. 358

## PYROFILM MAKES IT:

## High Voltage



- Four Sizes
- High Voltage Up to 20,000 Volts
- Resistance Range 50:2-300M
- Power Range Up To 2 Watts
- Temp. Coefficient $0 \pm 250 p$ P
- Radial Lead, Flal Protil



## Open-Frame Switchers!


LH's low-cost, single- and multi-ple-output Tiny-MITE switchers are perfect choices for high-volume OEM computer and terminal applications. TM-11 packs 100 wats it a $9.5^{\prime \prime} \mathrm{L} \times 4^{\prime \prime} \mathrm{W}$ 人 $2.5^{\prime \prime} \mathrm{H}$ package. TM- 23 packs 150 wars in- $811 x$ $5^{\prime \prime} \mathrm{W} \times 2.8^{\prime \prime}$ d. पM 34 ( 9045175 watts in $13.0^{\circ} \mathrm{L} \times 6^{\prime \prime} \mathrm{Y} .235^{\prime \prime} \mathrm{H}$.

- Output veltage - 1 M 11 : 5V $\pm 5 \%$ Noow. 23 main outputest -6820 amps . Various yolrages en 2nd and 3rd oftpats 1 M( 34 )rain output: 5 V -380 20 ampe. Various voltageser 2 nd 3 rid, and 4th outputs.
- Widerinpryariation - 92-130 or $180-268$ VAC, $47-450 \mathrm{~Hz}$.
- Ripprgand noise - $1 \%$ or 50 mv poaktopeak.
- Efficiency - TM-11: 75\%, TM23 and TM-34: 70\% nominal.
- Line regulation - $0.4 \%$ over entire input range.
- Load regulation - $0.4 \%$ from no load to full load.
- Response time - $200 \mu \mathrm{sec}$ to $1 \%$ after $25 \%$ load change @ 5 amp/ $\mu \mathrm{sec}$.
- Over-voltage protection - standard on main output.


## World's

 largest
## switcher

 manufacturer!LH Research produces the industry's broadest line of single- and multiple-output switchers. And nobody packs more power in smaller packages or offers more features. 1 through 7 outputs, up to $2.26 \mathrm{w} /$ in. $^{3}{ }^{3}$, $80 \%$ efficiency, and 2 -year guarantee. Less than $65 \$ / \mathbf{w}$ in quantity.

LH RESEARCH, INC.
1821 Langley Avenue. Invine, CA 92714

## COMPONENTS

Solid-state time delays preset at the factory


TDR Electronics, Foot of John St., Lowell, MA 01852. Gil Rogers (617) 459-0151. \$18.04 up; 2 to 4 wks.

TDR solid-state time-delay modules, factory preset, provide repeatable time delays from 1 to 300 s . The modules operate from 10 to 250 V ac or dc. Some types are available with total isolation between control and load circuits. The devices operate from -55 to 85 C and timing variation is $\pm 10 \%$ over the voltage and temperature range. Repeatability is $\pm 3 \%$ at fixed operating conditions and reset time is 500 ms . Off-state resistance is $20 \mathrm{k} \Omega$, minimum.

CIRCLE NO. 359

## Capacitors boast better rf characteristics



Republic Electronics, 176 E. 7th St., Paterson, NJ 07524. George Wolter (201) 279-0300. $\$ 0.43$ to $\$ 0.88$ (OEM); 8 to 10 wks .

A new ceramic formulation provides rf capacitors, Type 013Q, with a Q greater than 10,000 at 1 MHz , a QC product greater than 100,000 at 30 MHz and the ability to pass 4 A at 250 MHz with no power loss. The units are available in sizes from 0.1 to 1000 pF with voltage ratings from 100 to 500 V dc and a TC of $90 \pm 20 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$. Size is $0.110 \mathrm{in} .^{2}$ by 0.1 in . thick. Chips or five standard-lead types are available.

CIRCIE NO. 360

Panel pot has plastic resistance element


Allen-Bradley, 1201 S. Second St., Milwaukee, WI 53204. (414) 671-2000.

Conductive-plastic resistance elements and signal-level circuit switches are now offered in the MOD POT series of panel potentiometers. The conduc-tive-plastic, unitized resistance element affords smooth rotation and low turning torque. Also, contact-resistance variation is less than $0.2 \%$ and roll-on/roll-off, less than $0.25 \%$. The switch is tested with current levels as low as 15 mA and an open-circuit voltage of 5 V . Resistance range and tolerances are $100 \Omega$ to $1 \mathrm{M} \Omega, \pm 10 \%$ or $\pm 20 \%$. Linear, modified-log clockwise and counter-clockwise resistance tapers are available.

CIRCLE NO. 361

Drop out time delayed by solid-state timer


Artisan Electronics, 5 Eastman Rd., Parsippany, NJ 07054. Alan Seman (201) 887-7100. \$5; 2 to 3 wks.

The Model 437D solid-state timer delays the drop-out time of a relay when wired in series with a 48 -V-dc relay coil. The relay picks up in the normal time. With wire leads and solder lugs, the timer's square package is easily mounted. Operating on 30 to 60 V dc at 10 to 500 mA , the timer's delay can be preset from 10 ms to 600 s .

CIRCLE NO. 362

## Metal-film resistors meet MIL spec

International Components, 105 Maxess Rd., Melville, NY 11746. Mel Karasik (516) 29s-1500. \$36/M (5000 qty); stock to 12 wks.

The MF-25 1/4-W evaporated metalfilm resistors have a special epoxy coating for protection in severe environments. Characteristics meet MIL-R-22684 equivalent to the RN-55. Resistance ranges from $20 \Omega$ to $350 \mathrm{k} \Omega 2$ in standard $1 \%$ values. The tolerance is $\pm 1 \%$ with a TC of $100 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$. Packaging is on tape and reel, magazine pack, bulk pack or cut and formed to customer requirements.

CIRCLE NO. 363

## Ultramini fan weighs only 1.7 oz



Micronel, 8 Kane Industrial Park, Hudson, MA 01749. S.W. Linko (617) 568-8542.

With an output of 10 cfm in free air, the Model V60CL fan has an impeller diameter of 2.3 in . and weighs only 1.7 oz. It features a multivane impeller driven by a high-efficiency ac motor. CIRCLE NO. 364

## Multiposition rotary switch mounts on PC

Janco, 3111 Winona Ave., Burbank, CA 91504. (213) 845-7473.

Printed-circuit rotary switches, Series 2481, are for direct circuit-board or flex-circuit mounting use. These compact switches are 1 in . in diameter and available in $8,10,12,16,20,22$ and 24 positions. The shaft must be pulled in order to exit or enter certain predesignated positions. Secondary circuits are made or completed when the shaft is pulled. Completely enclosed and explosion-proof, the switches offer BCD output and meet or exceed MIL-S-3786 and/or MIL-S-22710.

CIRCLE NO. 365

## Micromini capacitors can mount on any side

Corning Electronics, Corning, NY 14830. (607) 974-9000.

A line of microminiature solid-tantalum capacitors with a unique coplanar design allows them to be mounted on any of their four sides. Called "Minichip" capacitors and desig-
nated the MK series, they are available in five case sizes and have electrical values ranging from $0.1 \mu \mathrm{~F}$ at 50 V to $47 \mu \mathrm{~F}$ at 10 V . Special ratings are available. The components have high resistance to mechanical shock, a low DF/ESR rating and good electrical stability. They can be used with automatic handling equipment and a vibratory bowl feed.

CIRCLE NO. 36
366


Miniature magnetic latching relay (MPCL) $0.6^{\prime \prime}$ PC board spacing-smallest in the industry. Two completely isolated coils designed for continuous duty with identical low power consumption, eliminating polarity reversal. Long life assured by patented hingeless armature construction. High armature holding force in latch position accompanied with simple permanent magnet circuit.


## Low-profile DIP relay (DR)

Lowest profile DIP relay available (only $0.38^{\prime \prime}$ high). Most sensitive DPDT relay-200 MW max. at pull-in, with 1 amp 28 VDC switching and 5 amp carrying capacity. Fits standard 16 -pin IC sockets, and its footprints are interchangeable with other e/m DIP relays. Costs less than 2 SPDT equivalent reeds.
New relays to meet customers' ever increasing demands. Write for literature to Gould. Inc., Controls Division. 131 Relay Road, Plantsville, CT 06479. Telephone (203) 621-6771.


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MODULES \& SUBASSEMBLIES
Synchro to digital unit fits into single module


Natel Engineering, 8954 Mason Ave., Canoga Park, CA 91306. Ed Berman (213) 882-9620. \$1195; stock to 6 wks.

Two-speed 20 -bit synchro to digital ( $\mathrm{s} / \mathrm{d}$ ) conversion is available in a single standard module. Model 2SD412 is 2.6 $\times 3.6 \times 0.8$ in., weighs less than 10 oz and is available with both TTL and LPTTL outputs at power consumptions of 3.5 and 2 W , respectively. Both binary and nonbinary ratios between the fine and coarse synchros are offered. Accuracy is $0.001^{\circ}$ at a speed ratio of $36: 1$. Standard inputs are 11.8 , 26 or $90-\mathrm{V}$ rms. The $400-\mathrm{Hz}$ reference can either be 26 or $115-\mathrm{V}$ rms.

CIRCLE NO. 367

## Light pens easily mate to CRT system



Information Control, 9610 Bellanca Ave., Los Angeles, CA 90045. Ron Hoover (218) 641-8520. \$450/\$475.

Two models of light pens, LP-210 and 211, simplify system interfacing for interactive CRT systems because of their high sensitivity and fast response. Sensitivity (with average brightness, P-31 phosphor, $20-\mathrm{mil}$ spot, 60 Hz ) is 2.0 foot-lamberts for both models. Response time is less than 300 ns. Spectral response is from 4200 to $11,000 \AA$; minimum vector speed is 20 $\mathrm{cm} / \mathrm{ms}$; minimum input separation is $20 \mu \mathrm{~s}$. The LP- 210 is activated by touching the operator's index finger to the pen's barrel. Model LP-211 has an actuation tip, enabled by pushing against the CRT faceplate.

## Op amps operate in harsh environments

Teledyne Philbrick, Allied Dr. at Route 128, Dedham, MA 02026. Frank Goodenough (617) 329-1600. \$62 to \$158 (unit qty.); stock.

Four high-performance militarygrade op amps, Models 1036, 1036-20, 1037 and 1037-20 can operate in harsh industrial and military environments from -25 to +85 C . The 1036/1036-20 output swings $\pm 10 \mathrm{~V}$ at $\pm 50 \mathrm{~mA}$ and its offset voltage ( $\mathrm{E}_{0 \mathrm{~s}}$ ) tempco is $\pm 10$ $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$. The output of the $1037 / 1037-20$ (a FET input device) swings $\pm 140 \mathrm{~V}$ at $\pm 10 \mathrm{~mA}$. Its initial $\mathrm{E}_{0 \text { s }}$ of $\pm 2 \mathrm{mV}$ eliminates any need for a trim potentiometer. The devices use hi-rel MIL components and meet environmental conditions of MIL-STD-202 and are processed to MIL-STD-883.

CIRCLE NO. 369

## A/d converter resolves 16 bits with 500 mW

Phoenix Data, 3384 W. Osborn Rd., Phoenix, AZ 85017. Srini Iyer (602) 278-8528. \$725 up; 8 to 12 wks.

The Model ADC 2000/2100 a/d converters have a resolution of 1 part in 65,535 ( 16 binary bits) with a total power consumption of only 500 mW . Accuracy is $\pm 0.004 \%$ and linearity is $\pm 0.002 \%$. All output and control signals are CMOS/TTL-LS compatible. The Model 2000 size is $5 \times 4.5 \times 0.5$ in. on a PC card and the Model 2100 is encased in $3 \times 4 \times 0.4 \mathrm{in}$.

CIRCLE NO. 370

## DIP clock provides ECL output

Vectron Labs, 166 Glover Ave., Norwalk, CT 06850. Larry Jawitz (203) 853-4433. \$95; 4 to 10 wks.

The C0-633 DIP-compatible crystal oscillator operates from -5.2 V dc and provides a stable ECL-compatible output. Frequency is in the 5 to $125-\mathrm{MHz}$ range. Initial accuracy is $\pm 50 \mathrm{ppm}$ for the CO-633A while that of the CO-633B is $\pm 10 \mathrm{ppm}$. Stability of each type is $\pm 25$ ppm from 0 to +70 C. Package size of the 14 -pin unit is $0.5 \times 0.8 \times 0.5 \mathrm{in}$. Options include the MIL-range CO-633-2 with stability of $\pm 50 \mathrm{ppm}$ from -55 to +125 C and the CO-633-3 that provides $\pm 3 \mathrm{ppm}$ from 0 to +50 C.

CIRCLE NO. 371

## Fixed-head memories operate on 48 V dc

Vermont Research, Precision Park, North Springfield, VT 05150. E.W. Hinkley (802) 886-2256. See text.

Fixed head memory, Model 4016-DC, operates on 48-V-dc input power for use in telecommunications systems where power interruptions cannot be tolerated. The device has a storage capacity of $37.9-\mathrm{Mbits}$ and access time of 8.5 ms . The memory occupies 12.25 vertical in. on a 19 -in. rack. All electronics, power conversion and controls are outside the media enclosure and easily accessible for maintenance. OEM price at $100 /$ year is $\$ 12,370$ for ac units and $\$ 13,390$ for de units at 4.64 -Mbyte capacity.

CIRCLE NO. 372

## Active filters enhance bio-medical signals

Frequency Devices, 25 Locust St., Haverhill, MA 01830. W. Morse (617) 374-0761. \$120; 2 to 4 wks.

Two types of active filters have frequency and transient responses that enhance bio-medical electronic signals. Model 7438 is a low-pass filter which provides 40 dB of attenuation at frequencies $15.5 \%$ above the cutoff frequency. Model 7439 high-pass filter attenuates all frequencies $15.5 \%$ below the defined cutoff, rolling at the same rate as the 7438 . Used in cascade, the two devices provide a flat-bandpass filter with steep skirts. Cutoff frequencies between 1 and 1000 Hz are available.

CIRCLE NO. 373

## Cable-matching amp spans dc to 1 MHz

Ross Engineering, 559 Westchester Dr., Campbell, CA 95008. (408) 377-4621. $\$ 800$ to $\$ 1200 ; 4$ to 8 wks.

A battery or ac operated, dc to 1 MHz , wideband, matching amplifier and 50 $\Omega$ cable driver acts as a buffer stage. Its 1 to $10-\mathrm{M} \Omega$ input impedance matches high-impedance output to a low-impedance load. The maximum output voltage ranges up to $12-\mathrm{V} \mathrm{pk}$ with input voltages from 12 to 1000 V pk and to $450-\mathrm{kV}$ pk with a proper voltage divider. The device has a builtin battery charger.

CIRCLE NO. 374

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## Use socket cards

 for probing and testing

Probe-Rite, 2725 Lafayette St., Santa Clara, CA 95050. (408) 249-1255. \$25 (24 qty); stock to 3 wks.

Two probing socket cards, P48-VI and P70-III, for use with manual probers and hand-plugged automatic testers, have a $0.1-\mathrm{in}$. spaced pattern. The boards include a programming area, edge connector, straight-through wiring, universal numbering system and standardized card dimensions for easy retrofit to most probing machines. A center-socket feature permits the die under investigation to be located directly below the optics when the card is placed in a probing machine.

CIRCLE NO. 375

## Mini terminal blocks

 handle solderless lugs

Square D Co., Milwaukee, WI 53201. (414) 332-2000.

Terminal blocks, Type GM3, are UL recognized and have solderless box lugs that accept \#10 to \#22 copper or aluminum wire. The blocks are made of rugged break-resistant thermoplastic polyester and are rated at 300 V . Designed for snap-in and snap-out mounting, their compact size accommodates up to 48 terminals/ ft of mounting track. Other features include a built-in wire strip gauge and marking area. Jumpers and separable connectors are available as accessories.

Knock-in wire mounts save space


Panduit, 17301 Ridgeland Ave., Tinley Park, IL 60477. (312) 532-1800. Free samples.

Low-profile, knock-in cable-tie mounts are available to secure wires on panels where space is at a premium. The mounts insert into pre-drilled holes and their plastic pins tapped down. The mount then sits flat on the panel surface. Three mounts are available in three sizes: KIMS-H500, 0.196 dia; KIMS-H430, 0.169 dia; KIMSH366, 0.144 dia. Material is $6 / 6$ nylon. Dimensions are $0.75 \times 0.5 \times 0.12 \mathrm{in}$. installed height.

CIRCLE NO. 377

## Logic jumpers provide up to 40 pins



Aries Electronics, P.O. Box 231, Frenchtown, NJ 08825. (201) 996-4096. See text.
Logic jumpers (DIPer) in 8 through 40-pin configurations are available using EIA color-coded, 26-gauge stranded wire. A female/male plug is wired on one wire end, and the other end can be terminated with a male plug or furnished with stripped and tinned wires for customer termination. Bifurcated contacts are gold plated for good electrical contact with either round, flat or square pins. The jumpers are available in any desired length from 6 to 10 in . Prices are $\$ 4.12$ to $\$ 26.10$ depending on number of pins and length of jumper.

CIRCLE NO. 378

## Pneumatic connector fits directly on PC boards



Linear Dynamics, P.O. Box 534, Wakefield, MA 01880. Don Denomme (617) 245-715\%.

Series 22 Pneutronics edge connector is compatible with most PC-card rack and motherboards. When mounted in conjunction with a standard PC-board edge connector, the Series 22 provides for a simultaneous plug-in of both electronic and pneumatic circuitry in a compatible modular package. Designed for quick disconnect operation, its construction requires no locking mechanism or fasteners to hold male and female connectors together under pressure or vacuum.

CIRCIE NO. 379

## IC-lead straightener adjustable to suit



Micro Electronic Systems, 8 Kevin Dr., Danbury, CT 06810. (203) 746-2525. \$199.50; stock.
A manual IC lead straightener, Fix-a-DIP, includes the proper number of adjustable spacers to accommodate IC widths of $0.3,0.4,0.5$ and 0.6 in ., 2 to 40 -pin DIPs. The unit can be factory adjusted to the width primarily required, and then later adjusted to other required widths. The device has a dual set of specially cut racks, spring loaded in the open position. A very badly bent IC lead must be hand straightened slightly, before it is inserted into the rack. Actuation of the soft plastic handle on the top of the unit straightens all the leads.

CIRCLE NO. 380

## Application notes



## Spectrum analysis

This may be the finest exposition on spectrum analysis that you're likely to see for a long time. It covers the fundamentals, the constraints, the math, the three fundamental approaches and the architecture in 48 well written, well illustrated pages. Rockland Systems, West Nyack, NY

CIRCLE NO. 381

## Ferrite toroids

Highlighted in a 24 -page catalog is a section containing design and application guidelines for the ferrite user. Included are formulas, material properties and coating information pertaining to rf, wideband and pulse transformers. Ferronics, East Rochester, NY

CIRCLE NO. 382

## Designing stamped parts

A 32-page guide outlines design techniques, for stamped parts, that cut costs and improve part quality. Dayton Rogers Mfg. Co., Minneapolis, MN

CIRCLE NO. 383

## Chip capacitors

"Understanding Chip Capacitors" covers all phases of ceramic-chip capacitor technology. The handbook is illustrated with performance graphs and comprehensive tables. Johanson Dielectrics, Burbank, CA

CIRCLE NO. 384

## Testing quartz crystals

"Crystal Testing Using a Spectrum Analyzer and Tracking Generator" discusses the characterization of quartz crystals for use in filters or oscillators using a conventional spectrum analyzer and tracking generator. Marconi Instruments, Northvale, NJ

CIRCLE NO. 385

## PROM programmer

An application note describes, for users of the SC/MP low cost development system (LCDS), a method of programming MM5204 and MM4204 PROMs that is both inexpensive and highly efficient. National Semiconductor, Santa Clara, CA

CIRCLE NO. 386

## Connector cross-reference

MIL-C-39012 Series SMA coaxial connectors and Sealectro part numbers are listed in a cross-reference. Rf Components Div., Sealectro, Mamaroneck, NY

CIRCLE NO. 387

# An $81 / 2$ inch Microprocessor controlled impact Printer for just $\$ 345^{*}$ 

 Now that's what we call practical!Laugh all the way to the bank, OEM's. With both matrix impact print head and built-in microprocessor controller, our DMTP-6uP is a budget printer in price You can practice, it's one of the greats You can print $80-96$ columns of both data
and text at a fast 110 cps . Turn out up to copies at once on regular $81 / 2$ inch roll paper. even on fan-fold forms and labels. Not only are all needle drivers and diagnostic routines included with the microprocessor, but you can choose the interface function you want - parallel ASCII, RS-232C/I-Loop, or You even get the baud rates from 110 to 1200 . economy of easily-replilion character life ribbon. - $\$ 345$ in 100 qts.: single units $\$ 472$


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New literature


## Power supplies

Over 3500 standard power-supply modules with complete electrical specifications, operating parameters, dimension charts, and prices are listed in a 60-page catalog. Abbott Transistor, Los Angeles, CA

CIRCLE NO. 388

## Relays

A 32-page catalog describes 878 models of reed, general-purpose, solidstate, power and sensitive relays, photocontrols and opto-isolators. The catalog contains specifications, schematics, outline drawings, and mounting templates, as well as complete ordering and pricing information. Sigma Instruments, Braintree, MA

CIRCLE NO. 389

## Indicator lamps

Described in a 16-page catalog are standard lenses, incandescent and neon indicators, cartridge hardware and lampholders. Each page covers a separate CM series: showing features, dimensional drawings, electrical and mechanical specifications. Chicago Miniature Lamp Works, Chicago, IL

CIRCLE NO. 390

## Plotter points

A four-page brochure describes plotter-point replacements for digital plotters. Koh-I-Noor, Bloomsburg, NJ

## Trimming pots

Technical specifications on subminiature trimming potentiometers are given in a four-page bulletin. Murata, Marietta, GA

CIRCLE NO. 392

## Instrumentation amplifiers

The theory of operation for both the Model 176 and 178 encapsulated instrumentation amplifiers is covered in a six-page brochure. The brochure also discusses common-mode rejection and frequency response. Calex Mfg., Pleasant Hill, CA

CIRCLE NO. 393

## PROM programmers

A 48-page guide describes PROM programmers, microprocessor system analyzers, microprocessor systems and support hardware, and design courses and seminars. Pro-Log, Monterey, CA

CIRCLE NO. 394

## Accelerometers

Photos, specifications and descriptions cover piezoelectric accelerometers. BBN Instruments, Cambridge, MA

CIRCLE NO. 395

## Data-acquisition system

The RECON remote data-acquisition and control system is described in a brochure. Photographs show typical installations for pump-station control, environmental monitoring, electricpower distribution, mining and energy management. Sangamo-Weston, Sarasota, FL

CIRCIE NO. 396

## Electronic enclosures

An eight-page brochure describes Accent vertical cabinets. Scientific-Atlanta, Optima Div., Tucker, GA

CIRCLE NO. 397

## Low-power Schottky TTL

A six-page folder covers 165 lowpower Schottky-TTL circuits. Included are operating-life-test data and a discussion of the company's Product Enhancement Program (PEP). Texas Instruments, Dallas, TX

CIRCLE NO. 398

## Connectors

A 64-page catalog features microminiature connectors and interconnect systems. Performance data, line and cutaway drawings and termination configurations are included. ITT Cannon Electric, Santa Ana, CA

CIRCLE NO. 399

## Encoders

Application information and electrical specifications for high resolution, absolute and incremental optical shaft-position encoders are given in a 20-page catalog. Litton Encoder Div., Chatsworth CA

CIRCLE NO. 403

## IPC technical reports

"How to Avoid Metallic Growth Problems on Electronic Hardware" discusses the problem in depth, reports on an industry survey on electromigration problems conducted by the IPC, and provides a series of positive actions that may be taken to inhibit or reduce the occurrence of this phenomena. Another report "Additive Process Evaluation" defines the results of a cooperative testing program in which 140 panels were submitted by IPC member companies for testing. The cost is $\$ 5$ per copy. IPC, 1717 Howard St., Evanston, IL 60202

CIRCLE NO. 408

## CMOS microprocessors

Features, descriptions, functional diagrams and instruction summary for the series 1800 CMOS $\mu$ Ps are given in a 20-page catalog. Hughes Solid State Products Div., Newport Beach, CA

CIRCLE NO. 409

## Disc drives

Design features, specifications and interface information on the 3300 series disc drives are given in an eightpage bulletin. Okidata, Mount Laurel, NJ

CIRCLE NO. 410

## Slide, rocker switches

A 25-page slide-and-rocker-switch catalog explains the company's new numbering system, switch-material specifications, UL and CSA ratings, and options. Schematics and PC layouts are included for each switch and the bulletin contains examples of custom designs and suggested wiring diagrams. Stackpole, Farmville, VA

CIRCLE NO. 404

## Voltage starters

Solid-state reduced-voltage starters are featured in a four-page brochure. Allen-Bradley, Milwaukee, WI

CIRCLE NO. 405

## Switchers

Specifications, outline dimensions, and single-unit prices of single and multiple-output switches are given in an eight-page catalog. Trio Labs, Plainview, NY

CIRCLE NO. 406

## Dedicated system

Two brochures describe the HP 2026 System. The 12 -page overview brochure describes system features, and gives a list of peripherals and usersupport programs. The 40 -page price/configuration guide includes information on warranty, base-system equipment and options. Hewlett-Packard, Palo Alto, CA

CIRCLE NO. 407

# IEEE Introduced a new interface standard. National Instruments introduced a new standard interface. 



IEEE introduced the new standard 4881975 General Purpose Interface Bus: and its popularity continues to grow

National Instruments introduced a complete package that allows PDP-11 users the versatility of using the growing number of instruments that comply with the IEEE 488-1975 standard bus. Now. PDP-11 users can configure complete computer control measurement systems, at low cost, and with commercially available instruments.

Our complete package includes the GPIB11-1 plug-in board. completely developed software, cables and connectors - everything you need to interface with 488-1975 compatible instruments. We offer off-the-shelf
delivery and thorough applications assistance. And we back your GPIB11-1 with in-depth service, testing and a one year guarantee.

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## Across the desk

(continued from page 12)
I support your viewpoint on "Professionals" $100 \%$. The lack of professionalism, as it might be called, is not only a sore thumb for the engineering community, but a sore thumb for most cross-sections of our society.

Ron Stier
Belden Corp.
P.O. Box 1327

Richmond, IN 47374

Don't let them put you down! Your laetrile editorial strikes at the heart of the biggest issue facing all of us, a government that can tell us the best things to buy, the proper foods to stuff our mouths with, the best way to be treated for disease-and people who collectively know little more about the real facts than any of us on the outside. Why can they do this? Because we let them. Why do they do it? Because they claim through a "feeling" for humanity that they are doing what is "best."

How can your scientific and logical readers simultaneously demand freedom to study, to investigate, to work and to live without interference and then claim that the Government must "protect" those "others" who are not as capable? Logical opinions do not necessarily come from people dealing with logic. When some of your more vocal readers can distinguish between grades of freedom and how they can be distributed between us and the unfortunate and misinformed, then publish their letters-and not before.

Stanley A. Fierston
7 Pickwick Road
Marblehead, MA 01945

Most people missed the most important point of the laetrile controversy. The issue is individual freedom, not the effectiveness or ineffectiveness of laetrile.

Each of us owns himself. Governments do not own us. If I want to shoot mayonnaise into my veins, that is my business and not any government's (by the way, some dumb kid did shoot mayonnaise when he couldn't get heroin and he promptly died). As far as I know, mayonnaise is not yet a controlled substance.

I don't care if laetrile is dangerous or safe, effective or ineffective, or some combination thereof. If I want to use laetrile, I will use it. Governments have no right to prevent my making it or buying it from someone who does make
it. I am an adult human being. I am responsible for earning my own living and keeping myself properly nourished, etc. Am I not also capable of determining or hiring someone to determine the safety, effectiveness, cost desirability, etc., of things that I wish to obtain and use?

I am an individual. I shall continue to do as I please while I observe the rights of others. I shall not knowingly violate anyone else's rights (property rights, human rights, etc.). Governments are hereby put on notice: Go jump in the lake!

> David Michael Myers Chief Engineer

Autocybernism Unlimited
Hughesville, MD 20637

I notice that two respondents who disagreed with your laetrile editorial mentioned thalidomide. I think that story needs to be told over and over until everyone comprehends what really happened and the implications.

Based on prescribed laboratory tests, many nations approved the manufacture and sale of this potent tranquilizer. The United States did not. Subsequently, in Europe and elsewhere, a rash of severely deformed children was traced to the use of this drug by their pregnant mothers. The United States was largely spared this horrible tragedy. The Food \& Drug Administration selected one of its top women "researchers," and in a public ceremony presented her an award for having prevented the distribution of this drug in the U.S.

But why did this entrenched bureaucracy fight against thalidomide? Was it inadequate testing, a lack of effectiveness, or, as proved to be the case, very undesirable side effectsitems the FDA is supposed to monitor? Hell, no! the decision was based on protecting the status quo, something the gone-to-seed bureaucracy is superb at. There already were approved drugs on the market to do the same job!

Raymond D. Musick Western Electric Co. 1111 Woods Mill Rd. Ballwin, MO 63011

Professionalism in the medicine business is nothing much more than a public relations technique to ensure a high income for the average physician and to protect him against the consequences of his own stupidity and arrogance.

Stupidity and arrogance? Those are the words.

Until about 1900, the chances of anyone being improved by a visit to a physician in the United States were about $50 \%$. Lower in most other piaces. Even today, about $20 \%$ of hospital cases are due to iatrogenic (physiciancaused) problems, and a large additional number to unnecessary, and frequently harmful, medical procedures and surgery, according to medical profession sources. Also, the various surveys conducted by the AMA show 10 to $20 \%$ of practicing physicians to be utterly incompetent or too drunk to practice, and perhaps $50 \%$ to be relatively incompetent.

Other places may be worse. About 100 years ago, for example, at the University of Gottingen Hospital, there was a period of three years during which not a single live mother or baby came out of the maternity wards. All were killed by puerperal fever induced by the physicians, physicians too arrogant even to consider the possibility that Semmelweis might have been right.

The medicine men will either ignore me, attack me on irrelevant grounds, or claim that things are different now. But of course they are not. The tradition of Gottingen lives on.

The physician of today is a body repairman, who hasn't a clue as to how the body works. He knows almost nothing of physics, damn little of microchemistry, a little bit about the essential chemical reactions of the body, and absolutely nothing about the internal workings of the brain. "Reputable" psychiatrists (MDs all) still go in for the gruesome procedures of electroshock "therapy," which is on a par with attempting to repair a large computer complex by passing lightning bolts through the power supplies.

But the tradition of medical infallibility lives on, and it is still almost impossible to get a physician to stand up in court and testify that a fraternity member goofed.

And the future will be no better. Go to any reputable university. Go to the elementary physics class and pick out the students who are failing. Give or take about $10 \%$, these are the pre-med students, your physicians of tomorrow.

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Dr. Yale Jay Lubkin
Director of Engineering
Ben Franklin Industries, Ltd.
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(continued on page 206)


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## Across the desk

(continued from page 204)

Laetrile should be judged on the question of whether it is helpful or harmful. There appears to be more evidence than many are aware of that it is helpful and little, if any, evidence that in normal doses it is harmful. The medical community itself is very divided on this subject. However, many doctors are using and supporting it. Even if laetrile is proved to be just another placebo (many of which have been used effectively by the medical profession for years), there is no reason to ban it. We cannot overlook the psychological cure of diseases and the relief of pain-in fact, this approach should be researched and expanded. Who cares how a person is helped as long as he is really helped?

Also, much evidence exists as to the dangers of smoking-but where is the big movement to ban it? The tobacco industry is, of course, much too wealthy and powerful to let that happen. And as we all must admit, it is the wealthy, hence powerful, groups that direct our government, not good old common sense and "what's best for the people."

As to the medical profession, reader Brody (ED No. 19, Sept. 13, 1977, pp. 188-190) seems to exemplify the members whose attitudes have resulted in a lack of faith in the medical community. We are aware of the countless number of unnecessary operations performed yearly, the fraud in Medicare programs, the ever increasing interest in more money rather than in the patients' health, and in many cases outright incompetence.

Then we read such statements as "I most respect the medical profession for looking out for \#1" and "I'd rather be a rotten S.O.B (don't worry Doc-you are) with a good, steady income and some solid security than that really hard working Nice Guy who's out of a job..." Where's the RESPECT Dr. Brody talks about? Professionalism certainly does mean respect and dedication to the profession, not to money! Certainly money is important and EEs should fight for a much better deal. But having a high income certainly does not make you a professional. Nor does a high formal education such as a Ph.D make you honest and trustworthy.

I became an EE and remain one because I love the profession. I spend a great deal of my own time and money
studying and trying to improve my knowledge and ability. There are many other jobs where I could make more money and have greater security, but I stay in electronics because it is more than just a job to me. It is my profession and even hobby-my second favorite thing, if you will (after all, I was a male before I was an EE).

Jerry Petrel
Electronics Design Engineer
3208 Navajo Way
Nevada Test Site
Las Vegas, NV 89108

It occurred to me in reading your second laetrile editorial (ED No. 19, Sept. 13, 1977, p. 51) that most people who agree with your position would not bother to write.

I am an electronics designer who gets his laetrile by eating about five peach or plum pits every day-to hell with the FDA.

Your outspokenness is very much appreciated-keep up the good work!

Cznko Funk
Project Engineer
Fitch Creations, Inc.
North Greensboro St.
Carrboro, NC 27510

Three cheers for your September 13 stand on professionalism. If enough specialists in all fields cannot become personally responsible to the public at large for their own conduct and for their own set of priorities, we are all probably doomed.

Who can police the professional but other similarly trained professionals? An everyone-for-himself-to-hell-with-everyone-else attitude destroys the mutual trust on which modern (technologically enslaved) society depends.
F.L. Walker

Principal Engineer
Raytheon Data Systems
1415 Boston-Providence Turnpike Norwood, MA 02062

Judging from the negative replies in the September 13 issue, your July 19 editorial on professionals using the current "nonprofessional topic"-lae-trile-was sorely needed.

You picked a good selection of letters, as they covered just about every irrationality and prejudice associated with laetrile. As you stated in your rebuttal, the negative replies-and some of the positive ones-missed the basic idea of the editorial: A pro-
fessional who does not keep an open mind is not professional-or worse. Unfortunately, the medical-industrial complex is the most unprofessional group around-except for big government.

Keep on reminding us of our professionalism.

Peter E. Sluka
Western Electric
Naperville, IL 60540

## By the way...

The computer-program illustration that appeared in ED No. 21, Oct. 11, 1977, p. 36 was reproduced from an article that appeared in the May, 1977, issue of BYTE magazine.

Misplaced Caption Dept.


When the recession hit and our funds for development were cut off, we could understand that. But now, the situation is actually worse

Sorry. That's a sculpture of Diadoumenos (Roman copy) by Polycleitos, which stands in the Athens National Muscum.

## Anonymous con

Occasionally, Mr. Rostky is so bright he's out of sight. He reminds me of our QC Dept.-great at pointing out problems they do not have to solve.

Name withheld Marconi Instruments Ltd.
St. Albans, Herts, UK

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[^3]:    Frank Deverse, President. International Microcircuits, 3004 Lawrence Expressway, Santa Clara, CA 95051.

[^4]:    Satya Pal Asija, P. E., Senior Applications Analyst, International Defense Systems Div., Sperry Univac Corp., St. Paul, MN 55117

[^5]:    Thomas M. Reme, Product Engineer, Alpha Wire Corp., 711 Lidgerwood Ave., Elizabeth, NJ 07207.

[^6]:    2. A j-k flip-flop following the comparators allows the circuit to be clocked to synchronize input data. In addition to hysteresis from the comparators, clocking helps reduce noise rejection even further.
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