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This month's cover picture shows part of the internal structure of an image intensifier made by Cathodeon. (Photographer Paul Brierley)

\section*{In our next issue}
(publication date October 15)
Model railway control system, using different d.c. levels, provides control of speed, locomotive whistle and coach interior illumination. Multi-flash trigger unit initiates flashes a equal intervals from milliseconds to seconds for sequence photography.

October 1973
Volume 79 Number 1456

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In our July issue we published a letter bewailing the lack of a "figure of goodness" for f.m. tuners. In the August issue J. R. Stuart's article on amplifier design told of the attempt to find a co-relation between objection measurements and subjective tests of harmonic distortion. These passages, and no doubt the unexpressed thoughts of many audiophiles, indicate a yearning for that mysterious formula, an objective standard of quality in reproduced sound. In its absence we must make do with subjective assessments. Fortunately, for those in mental distress, there are plenty of people around who will give you a subjective assessment at the drop of a hat. There are those who can tell the difference between "transistor sound" and "valve sound", those who can distinguish between a high quality Class B amplifier and a high quality Class A one, and above all there are the reviewers of equipment in the audio magazines.

One presumes that these people (especially the reviewers, who publish their conclusions) have a highly discriminating sense of hearing. Perhaps they were born with it, but at any rate we are told that whatever the initial state of the faculty it can be developed by training - the ear can be educated. But what exactly is the nature of the training, what are the criteria for an educated ear, what levels of aural education are attainable, how do we know what level a particular assessor has reached. and to what extent can we rely on the judgments resulting from the training? In fact we know very little about these qualifications. It appears that the training is self-administered and consists largely of constant listening, straining to distinguish minute differences of sound quality, development of a good aural memory, and familiarization with the aural correlates of engineering measurements such as distortion. From advertisements in which items of audio equipment are endorsed by famous musicians we are led to believe that these powers of assessment are normal attributes of musically trained people - and, perhaps conversely, that equipment reviewers and other assessors have musical ability of the kind possessed by professional musicians. But this is dangerous territory - a quagmire of subjectivism.

For those who feel lost in such considerations it is comforting to know that there is, at least, an objective standard of audio equipment performance. This is the group of specifications, DIN 45500 , issued by the German standards institution Deutscher (Industrie) Normenausschuss, which lays down in quantitative engineering terms the minimum performance requirements for domestic equipment that will permit it to be described as "high fidelity". Regrettably the British Standards Institution does nothing equivalent (only recommended methods of measurement), so those British audio manufacturers that are interested are making use of DIN 45500 . For the customer who does not have an educated ear but does have a highly developed sense of value for money it must be some satisfaction to know that what he has bought is, in a sense, sound of guaranteed minimum quality. The conformity to DIN 45500 and the figures contained in the specifications do not tell him the nature of this sound quality he has bought but, whatever it is, he does know that he has got it.

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\author{
by A. J. Ewins
}

This article describes the design and construction of the electronics of a \(3 \frac{1}{2}\)-digit, dual-polarity digital voltmeter for measuring alternating and direct voltages in the range 200 mV to 400 V full-scale. It is capable of an accuracy of \(\pm 0.05 \%\) of full-scale which in practice is limited by the precision of the input attenuator resistors and the accuracy of the a.c. rectifier circuit. A relatively inexpensive attenuator system is described which is capable of being adjusted to an accuracy of \(\pm 0.1 \%\) giving an overall accuracy to the d.c. ranges of \(\pm 0.1 \%\) of reading, \(\pm 0.05 \%\) of full-scale. For the a.c. ranges the accuracy is further limited by the rectifying circuit but nevertheless an accuracy of \(\pm 0.5 \%\) of reading, \(\pm 0.05 \%\) of full-scale is achieved over a frequency range of 30 Hz to 100 kHz . The fullscale reading of the d.v.m. is 1999 (with suitable positioning of the decimal point) and an overload indication is given for readings in excess of this value. In spite of the overload indication it is possible to interpret readings up to an indication equivalent to 2500 .
An additional feature of the d.v.m. is its ability to operate in either an automatic or a manual mode. In the auto. mode the d.v.m. continually samples the input voltage but can be made to "hold" and display the latest reading by depressing a sample/hold switch. In the manual mode the d.v.m. holds and displays the latest reading of the input voltage but can be made to further sample it by depressing the sample/hold switch, after which it again holds.
Details of the construction of the electronics on suitable circuit boards only is
given, it being left to the reader to devise a suitable cabinet construction.

\section*{Design principles}

The d.v.m. operates on the by now wellknown dual-slope integration method. Two previous designs presented in this journal \({ }^{1,2}\) operate on this principle and a detailed account of the theory was given in Waddington's article, "Digital Multimeter" \({ }^{1}\). However, for those readers who missed both previous articles and are unfamiliar with the technique a brief description of the principle will be given.

Single polarity. With reference to the block circuit diagram of Fig. 1, a capacitor \(C\), is charged (negatively) by connecting the positive input voltage, \(V_{i n}\), (via the electronic switch) to the resistor \(R\) for a known time. The resistor \(R\), capacitor \(C\) and the highgain operational amplifier form an integrating circuit with highly linear characteristics. At the end of this time the voltage on the capacitor will be \(V_{c}\), given by the expression:
\[
V_{c}=(1 / R C) \int_{0}^{t_{1}} V_{i n} \cdot \mathrm{~d} t=V_{i n} \cdot t_{1} / R C .
\]

The electronic switch now connects the resistor \(R\) to the accurately known negative reference voltage and the capacitor is discharged (positively) until \(V_{c}\) reaches the value it was before it was charged (usually , zero volts). Thus,
\[
\begin{aligned}
0=V_{c}-(1 / R C) \int_{0}^{t_{2}} & V_{r e f} \cdot \mathrm{~d} t \\
& =V_{c}-V_{r e f}, t_{2} / R C .
\end{aligned}
\]

Hence, \(V_{\text {in }} \cdot t_{1} / R C=V_{\text {ref }} \cdot t_{2} / R C\) and \(V_{\text {in }}=\) \(V_{\text {ref }} \cdot t_{2} / t_{1}\).

The value of \(t_{1}\) is determined by the frequency of the clock oscillator and the divide-by- 10 and 2000 counters. (The divide-byten counter has been introduced to ensure that the control logic operates at a speed at least ten times faster than the time represented by the least significant digit of the main, divide-by- 2000 counter.) At the beginning of the charging process these two counters are set to zero. The logic is so arranged that when the divide-by-2000 counter reaches a total count of 2000 (after 20,000 clock pulses, and hence after a time \(2 \times 10^{-5} \times 10^{4}=100 \mathrm{~ms}\) ) the charging process is stopped and the discharging process begun (by disconnecting \(R\) from \(V_{\text {in }}\) and connecting it to \(-V_{\text {ref }}\) ). The divide-by2000 counter now counts the number of clock pulses (divided by ten) received in the time it takes to discharge the capacitor to zero volts. The comparator, of Fig. 1, detects the zero voltage level and transmits a pulse to the control logic, commanding the transfer of the time indicated by the main counter to the display. Immediately after the transfer, the divide-by- 10 and 2000 counters are reset to zero and the resistor \(R\) reconnected to \(+V_{i n}\). A new measurement cycle then begins. (N.b., there is no need to reset the divide-by-10 and 2000 counters to zero at the end of the charging process because they automatically recycle at the end of this period.) Because the fullscale of the main counter is effectively a display of 2000 , the value of \(V_{\text {in }}\) may be read directly in terms of \(t_{2}\) if \(-V_{\text {ref }}\) is given a value of -2 volts, i.e. \(V_{i n}=t_{2} \cdot 2 \mathrm{~V} / 2000\) This is the case of the d.v.m. described in this article and the basic range is thus 2 volts.

The beauty of the dual-slope integration method lies in the fact that the long-term stability of the clock oscillator and the precise values of \(R\) and \(C\) are unimportant. Providing they are stable over the shortterm (not more than 200 ms for a full-scale reading) which may more than reasonably be expected, the accuracy of the system is theoretically only dependant upon the accuracy of the reference voltage.

As has already been mentioned, the divide-by-10 and 2000 counters are reset to zero immediately after the time \(t_{2}\) has been transferred to the display. This is unlike the
previous designs described in this magazine, which had a "dead period" at the end of the discharging process, whilst the main counter continued towards a total count of 2000 , before reconnecting the input to the integrator to \(V_{i n}\). The modification to this design allows readings in excess of full-scale (up to a limiting value discussed later) to be accurately interpreted and is one of the reasons why existing large-scale integrated circuits were not considered for the control and display logic. The main reason, at the time of construction, was one of cost and availability. The article, "Digital Panel Meter" \({ }^{2}\), published in this journal, amply illustrates the advantages to be obtained from using l.s.i.cs. However, while the use of individual logic circuits may be cheaper than the large-scale ones (if considerably more bulky) the author has taken advantage of the opportunity to "stamp" his own originality upon the design of d.v.m. circuits. Eventually, when the cost of 1.s.i.cs become really low, the designer will no longer be free to indulge his own whims and fancies upon the basic designs, but will have to be content to accept the designs of others.

The lack of a "dead period" in the operation of the d.v.m. measurement cycle does have one other advantage in the simplification of the switching circuitry when only one resistor is used in the integrator. If more than one resistor is used, say, one for \(+V_{\text {in }}\) and one for \(-V_{\text {ref }}\), a possible source of error is introduced due to the differing ageing characteristics between resistors.

Dual polarity. The principle of operation of a dual-slope integrating digital voltmeter so far described has only been for a single polarity type. For a dual-polarity type, some modification to the basic block diagram of Fig. 1 is necessary. Fig. 2 is a block diagram of the dual-polarity principle adopted by the author. The method of conversion of voltage into time is exactly as described before. However, the capacitor \(C\) will be charged negatively or positively as determined by the polarity of the input voltage. Thus two reference voltages, of equal magnitude but opposite sign, are needed to discharge the capacitor, positively or negatively as appropriate. Two comparators are also required to detect the direction of the
charge on the capacitor and hence select the appropriate reference voltage for discharging. Except for the addition of polarity detecting logic, the operation of the control and display logic remains unaltered.

Initially the capacitor \(C\) is discharged and the output from the integrator, \(V_{c}\), will satisfy the following condition; -2 mV \(<V_{c}<+2 \mathrm{mV}\). The output from comparator 1 will therefore be logical " 1 " and that of comparator 2, logical zero. Due to the inversion of the logical level at the output of comparator 2 by inverter 1, the input to the control logic and from the AND gate will be logical " 1 ", which is the same as for the single-polarity d.v.m. of Fig. 1. This is thus the beginning of the measurement cycle and the control logic connects the resistor, \(R\), via the electronic switch, to \(V_{i n}\). Providing that the magnitude of \(V_{\text {in }}\) is greater than zero, the capacitor \(C\) will be charged. If \(V_{\text {in }}\) is positive, \(C\) will be charged negatively and the output of comparator 1 will become logical zero. Comparator 2 remains unaltered at the logical zero level.

The output from the AND gate thus changes to logical zero. Similarly, if \(V_{\text {in }}\) is negative, \(C\) will be charged positively and the output of comparator 2 will become logical " 1 ", comparator 1 remaining at the logical " 1 " level. Once again, the output from the AND gate becomes logical zero. At the end of the timing period, the control logic commands the connection of one of the reference voltages to the input of the integrator. The reference voltage selected is dependant upon the logical states of the two comparators fed-back to the electronic switch. If the output from both comparators is logical zero, \(-V_{\text {ref }}\) will be selected and if they are both logical "1", \(+V_{\text {ref }}\) will be selected. The logic of the electronic switching circuit is so arranged that a logical zero at the output of either comparator inhibits the selection of \(+V_{\text {ref }}\) and a logical " 1 " at the output of either comparator inhibits the selection of \(-V_{r e f}\). Because of this it is impossible for both reference voltages to be selected simultaneously. Having connected the appropriate reference voltage to the integrator the capacitor is discharged until the voltage across it again satisfies the condition, \(-2 \mathrm{mV}<V_{c}<+2 \mathrm{mV}\). When this occurs, the output from the AND gate
again becomes logical " 1 ", commanding the control logic to transfer the time \(t_{2}\) from the main counter to the display, reset the divide-by- 10 and main counters to zero and reconnect the input of the integrator to \(V_{i n}\).

Polarity indication is achieved by detecting the logical states of the two comparators at the end of the initial timing period. If at the end of this time the outputs of both comparators are logical zeros (when \(V_{\text {in }}\) is positive) the \(J\) input to the \(J-K\) flip flop will be logical " 1 "' (due to the inverting action of inverter 2) and the \(K\) input will be logical zero. A pulse from the control logic, transmitted at the end of the timing period, transfers the logical " 1 " at the \(J\) input to the \(Q\) output, which in turn operates drivers in the display unit to indicate the positive sign. Similarly, if at the end of the timing period the outputs of both comparators are logical " 1 ", then the \(J\) and \(K\) inputs will be logical zero and " 1 ", respectively, and a logical " 1 " will be transferred to the \(\bar{Q}\) output. Thus the negative sign will be displayed. When the input voltage is zero, the outputs of both comparators remain unaltered, from their original conditions, at the end of the initial timing period. Both \(J\) and \(K\) inputs are thus logical zeros and the polarity sign indicated is that of the last input voltage greater than zero. In this manner, the sign indicated by the display remains illuminated until an input voltage of opposite polarity is applied. By shorting the \(V_{i n}\) input to earth, the polarity indication may be used to accurately set the zero reading of the d.v.m.

The -2 mV and +2 mV reference voltages applied to the negative inputs of comparators 1 and 2, respectively, are necessary in order to define clearly the discharged state of the integrating capacitor. It would not be possible to do this if both reference inputs were connected to earth. In practice it is sufficient to separate the reference levels of the comparators by such an amount as to guarantee the discharged state to be clearly defined, allowing for temperature drift of the differential offset voltages of the comparators. The fact that the two comparator reference levels are at different voltages in no way impairs the accuracy of the system. However, the closer the two reference levels, the easier it is to assess a true zero condition in the absence

Fig. 2. The system of Fig. I adapted to accept dual-polarity inputs.

of an input voltage, and 2 mV represents a resolution of less than the least significant digit displayed.

\section*{Basic d.v.m. circuit}

For convenience the basic circuitry of the d.v.m. has been split up into two. Fig. 3 is a detailed circuit diagram of all the analogue circuitry of the d.v.m. plus the electronic switching circuitry and the polarity logic. Fig. 4 is a detailed block circuit diagram of the control and display logic. It will be seen that the total circuitry uses only well known and easily obtainable, inexpensive, operational amplifiers, transistors and t.t.l. digital circuits.

Analogue circuit. Examining first of all the circuit of Fig. 3 it will be seen that \(Z_{2}\), which is fed with a constant current of about 5 mA by \(T r_{5}\) and its associated circuitry, provides a reference voltage of +5.6 volts. This large positive reference voltage is converted into a smaller positive one of 2 volts, with low output impedance, by \(I C_{1}\) and its associated circuitry. It is similarly converted into a -2 volt reference level, with low output impedance, by the inverting action of \(I C_{5}\) and its circuit. Variable resistors, \(R_{1}\) and \(R_{2}\), allow for a precise setting of these two reference levels. \(I C_{3}\) provides a buffer input for \(V_{i n}\), has unity gain and a low output impedance. The output from \(I C_{3}\) is voltage limited by zener diodes \(Z_{3}\) and \(Z_{4}\) for the protection of
the transistors used in the electronic switch. \(R_{3}\) allows the output from \(I C_{3}\) to be adjusted precisely to zero when the input voltage is zero. \(I C_{6}\) is the integrating operational amplifier and \(I C_{8}\) and \(I C_{9}\) are the comparators, 1 and 2 , respectively. \(T r_{1}\) is the transistor switch for \(+V_{r e f}, T r_{4}\) the switch for \(-V_{r e f}\) and transistors \(T r_{2}\) and \(T r_{3}\), in parallel, the switch for \(V_{i n}\). The operation of these transistor switches is controlled by \(I C_{2}\) and \(I C_{4}\) and the outputs from the two comparators via the diodes \(D_{1}\) to \(D_{8}\). The operation of these transistors as switches may be unfamiliar to some readers and will therefore be explained. Figs. \(5(\mathrm{a})\) and \(5(\mathrm{~b})\) illustrate, respectively, the conventional use of a transistor as a switch and the more unconventional mode as used in this d.v.m. In Fig. 5(a), when a sufficiently large current is applied to the base of the transistor, it saturates with a collector-emitter voltage of typically 200 mV , which varies little with varying supply volts, \(V_{s}\), providing the collector resistor is of a reasonably high value. This offset voltage is very high when compared with a resolution for the d.v.m. of 1 mV and makes the use of a transistor as a switch, in this mode, quite unsuitable. It is a curious fact, however, that when the same transistor is turned upside down, see Fig. 5(b), ( \(V_{\mathrm{s}}\) must not now exceed the \(V_{b e}\) breakdown voltage of the transistor) and a sufficiently large current is applied to the base of the transistor, it saturates, this time with an
emitter-collector voltage of only a few millivolts. Experiments with n-p-n and p-n-p transistors, types BCl 182 and BC 212 respectively, in the test circuits of Figs. 5(c) and \(5(\mathrm{~d})\) (which are effectively rearrangements of the circuit of Fig. 5(b), allowing for transistor types), yielded the results in Table 1.

Table 1
\begin{tabular}{ccc}
\hline \multirow{2}{*}{\begin{tabular}{c}
\(V_{i n}\) \\
(volts)
\end{tabular}} & \multicolumn{2}{c}{\(V_{c \theta}(\mathrm{mV})\)} \\
\cline { 2 - 3 } & BC182 & BC212 \\
\hline+2 & +3.3 & +2.1 \\
+1 & +1.5 & +1.3 \\
0 & -0.3 & +0.3 \\
-1 & -1.7 & -1.0 \\
-2 & -2.8 & -2.5 \\
\hline
\end{tabular}

The above two transistors were selected randomly and others of the same two types produced only slightly different results From the above results it was thought quite satisfactory to use a BC182, in this switching mode, for switching \(+V_{r e f}\) and a BC212 for switching \(-V_{r e f}\). The very small voltages dropped across the collector and emitter of the transistors are easily allowed for in the adjustment of the two reference levels.

Neither of the two transistors was thought ideal for switching the input voltage because, although the small voltages dropped across the transistors at \(V_{\text {in }}\) equals zero could be allowed for in the zero adjustment


Fig. 3. The analogue circuitry.
of the input op-amp, the relationship between \(V_{c e}\) and \(V_{i n}\) were not exactly linear. However, the author discovered that if both types of transistor were used (by connecting them back-to-back as in Fig. 5(e)) and were switched on simultaneously, a nearly ideal switch was produced. Using the same two transistors, as tested above, in the experimental circuit of Fig. 5(e) produced the results in Table 2.

Table 2
\begin{tabular}{cc}
\hline\(V_{i n}\) (volts) & \(V_{c \theta}(\mathrm{mV})\) \\
\hline+2 & +2.7 \\
+1 & +1.4 \\
0 & \(\sim 0\) \\
-1 & -1.3 \\
-2 & -2.7
\end{tabular}

The two transistors were thus acting together to produce a switch with very nearly zero offset voltage and an effective "on" resistance of \((2.7 \mathrm{mV} \times 20 \mathrm{k} \Omega) / 2 \mathrm{~V}=27 \Omega\). A parallel combination of a BCl 82 and a BC 212 is thus used as the switch for \(V_{i n}\). The above table of results is interesting in view of the fact that the two transistors used were not precisely matched, except for their values of \(V_{c e}\) at zero input voltage. The author, therefore, feels fairly confident that any pair of transistors, types BC182 and BC212, should function satisfactorily in this manner without the need for special matching.

(a)

(b)

(c)

(d)


Fig. 5. (a) Conventional and (b) inverted transistor switches. Test circuits, using n-p-n and \(p-n-p\) devices are shown in (c) and (d) and (e) in the final form using both types.


Fig. 6 illustrates the complete switching circuitry of the d.v.m. and, remembering that the outputs from the two comparators (IC \(C_{8}\) and \(I C_{9}\) ) and \(I C_{2}\) and \(I C_{4}\) are either + or -10 volts (approx.), operates as follows: At the onset of the measurement cycle (i.e. \(V_{\text {in }}\) connected to the input of the integrator) the voltage at the output of \(I C_{8}\) is +10 V ; that of \(I C_{9},-10 \mathrm{~V}\); that of \(I C_{2},-10 \mathrm{~V}\) and \(I C_{4},+10 \mathrm{~V}\). The voltage at the junction of the diodes, \(D_{1}, D_{2}\) and \(D_{3}\), is thus approx. -10 V and \(T r_{1}\) is OFF, the voltage at the junction of the diodes, \(D_{6}, D_{7}\) and \(D_{8}\), is approx. +10 V and \(T r_{4}\) is OFF; the voltage at the base of \(T r_{2}\) is negative of its collector and therefore it is ON , and the voltage at the base of \(\mathrm{Tr}_{3}\) is positive of its collector and therefore it is also ON . At the end of the timing period the voltages at the outputs of \(I C_{2}\) and \(I C_{4}\) reverse, to become +10 V and -10 V respectively, turning transistors \(T r_{2}\) and \(T r_{3}\) OFF. In the absence of any feedback from the outputs of the two comparators, transistors \(T r_{1}\) and \(T r_{4}\) would simultaneously be turned ON, shorting \(+V_{\text {ref }}\) and \(-V_{\text {ref }}\) together. However, with the connections as shown and assuming a positive input voltage, the output of comparator one \(\left(I C_{8}\right)\) will have become -10 V by the end of the timing period, the output of comparator two remaining -10 V . The junction of diodes \(D_{6}, D_{7}\) and \(D_{8}\) is thus free to swing in a negative direction at the command of \(I C_{4}\) and \(T r_{4}\) is turned ON connecting \(-V_{\text {ref }}\) to the input of the integrator. -10 V at the outputs of both comparators prevent \(T r_{1}\) from being turned ON . If the input voltage had been negative, the output of comparator two ( \(I C_{9}\) ) would have become +10 V by the end of the timing period, the output of comparator one remaining +10 V ; the junction of diodes \(D_{1}, D_{2}\) and \(D_{3}\) would then have been free to swing in a positive direction at the command of \(I C_{2}\) and \(T r_{1}\) would have been turned ON , connecting \(+V_{\text {ref }}\) to the input of the integrator. +10 V at the output of both comparators prevent \(T r_{4}\) from being turned ON . The diodes \(D_{1}, D_{4}, D_{5}\) and \(D_{8}\) prevent breakdown of the emitter-base junctions of the transistors \(T r_{1}\) to \(T r_{4}\), respectively, when they are held in their OFF states.

The switching currents feeding into the bases of the transistors \(T r_{1}\) to \(T r_{4}\) to turn them ON are of the order of \(300 \mu \mathrm{~A}\) to \(500 \mu \mathrm{~A}\), and it is because of this relatively heavy current that the output impedances of the sources of \(+V_{\text {ref }},-V_{\text {ref }}\) and \(V_{\text {in }}\) must be low. A higher output impedance would result in these voltage levels being altered in the presence of the switching currents, impairing the accuracy of the d.v.m.
Returning now to Fig. 3, it will be appreciated that the outputs from the two comparators (either + or -10 volts) are not compatible with t.t.1. digital circuit logic levels. (For the t.t.1. circuits, logical \(0 \equiv 0.2 \mathrm{~V}\) and logical \(1 \equiv 2.5\) to 5.0 V .)
Transistors \(T r_{6}\) to \(T r_{9}\) are included to achieve this conversion in voltage levels and to provide the inversion function of the inverters 1 and 2 of Fig. 2 and also to provide the function of the AND gate. Thus, when the output from \(/ C_{8}\) is +10 volts, the collector of \(T r_{9}\) is about 0.2 volts and hence the


Fig. 6. The complere analogue switching circuit.
\(J\) input of the \(J-K\) flip-flop is logical " 0 "; when it is -10 volts, the collector of \(T r_{9}\) and the \(J\) input are at about 5 volts, i.e. logical " 1 ". Similarly, when the output from \(I C_{9}\) is -10 volts, the collector of \(T r_{6}\) is about zero volts and hence the \(K\) input to the \(J\)-K flip-flop is logical " 0 "; when it is +10 volts, the collector of \(T r_{6}\) and the \(K\) input are at the logical "l" level. When the outputs from \(I C_{8}\) and \(I C_{9}\) are +10 volts and -10 volts respectively, the output to the control logic, \(V_{o}\) (the junction of the collector of \(\mathrm{Tr}_{7}\) and the emitter of \(\mathrm{Tr}_{8}\) ), is at the logical " 1 " level. If either \(I C_{8}\) is -10 volts or \(I C_{9}\) is +10 volts, the output, \(V_{o}\), becomes logical zero.
\(I C_{2}\) and \(I C_{4}\) are included for similar reasons to the transistors \(T r_{6}\) to \(T r_{9}\); they convert the t.t.l. logic level from \(\bar{C}\) to the necessary plus and minus 10 volt levels for the operation of the electronic switch circuitry. The positive input of \(I C_{2}\) and the negative input of \(l C_{4}\) are held at a voltage level of about +2 volts. A logical " 0 " at the \(\bar{C}\) output from the control logic thus produces voltage levels at the outputs of \(I C_{2}\) and \(I C_{4}\) of +10 V and -10 V , respectively. Logical " 1 " at the \(\bar{C}\) output produces voltage levels of -10 V and +10 V at the outputs of \(I C_{2}\) and \(I C_{4}\), respectively.

It has already been stated that the inclusion of zener diodes \(Z_{3}\) and \(Z_{4}\) is to limit the output from \(I C_{3}\) for the protection of the switching transistors. The necessity for this protection can best be understood by assuming that a voltage much greater in magnitude than 2 V , say +4 V , were applied to the input of \(I C_{3}\). Since it has unity gain, the voltage on the collectors of \(\mathrm{Tr}_{2}\) and \(\mathrm{Tr}_{3}\) would also be +4 V . Now, if the input of the integrator were connected to \(-V_{\text {ref }}\) then the voltage on the emitters of \(T r_{2}\) and \(T r_{3}\) would be -2 V and the emitter-base junc-
tion of \(T r_{2}\) would be in danger of breaking down, since \(V_{b e}\) for all four transistors is only rated at 5 volts. (N.b., a transistor with a reverse voltage connected across its emitter and collector can only withstand a voltage equivalent to its \(V_{b e}\) breakdown voltage plus the voltage across the forward biased collector-base diode.) Similar danger would be experienced if \(V_{i n}\) were -4 V and the input to the integrator connected to \(+V_{\text {ref }}(+2 \mathrm{~V})\); the emitter-collector junction of \(\mathrm{Tr}_{3}\) would then be threatened. The emitter-collector junctions of \(T r_{1}\) and \(T r_{4}\) could also be threatened if the input to the integrator were connected to \(V_{i,}\) when it was at a level of +4 V or -4 V , respectively. \(Z_{3}\) and \(Z_{4}\) limit the voltage on the collectors of \(T r_{2}\) and \(T r_{3}\) to approx. \(\pm 3.3\) volts so that the magnitude of the maximum possible voltage across the collector-emitter junctions of all four switching transistors is 5.3 volts.

The variable resistors \(R_{4}\) and \(R_{5}\), of Fig. 3, allow the short circuit and open circuit input offsets of the integrating op-amp \(I C_{6}\), to be set to zero.
The action of \(I C_{7}\), together with \(T r_{10}\) and its associated circuitry, allows the output of the integrator to be held at the zero volt level by shorting the integrating capacitor. This circuitry is used in conjunction with the auto/manual facility of the d.v.m.

Control logic. The inter-connections between the various t.t.l. digital circuits of the control and display logic are shown in the block diagram of Fig. 4. The second, third and fourth decade counters and the divide-by-two flip-flop form the basic divide-by2000 counter. \(A, B, C, D, E, F\) and \(G\) are all \(J-K\) flip-flops and are contained in four t.t.I. circuits, type SN7473. Not shown are the connections between the outputs of the divide-by-two flip-flop and the clock pulse, \(C_{p}\), input of flip-flop \(A\), and the output of the \(F\) flip-flop and the \(C_{p}\) input of flip-flop \(G\). The six inverters, which are buffer/ drivers with open collector outputs capable of sinking 40 mA , are all contained in the t.t.l. circuit, type SN7416. The four dualinput NAND gates are all contained in the t.t.l. circuit, type SN7400, and the four dualinput AND gates in the t.t.l. circuit, type SN7408. The clock circuit is made from a t.t.1. dual-Schmitt trigger circuit, type SN7413, together with a resistor and a capacitor. The circuit is shown in Fig. 7.

The binary outputs from the three decade counters of the main counter are transferred to suitable decoders via three quadruple latches, t.t.l. circuits type SN7475. In a similar manner, the outputs from the divide-by-two flip-flop and the \(A\) flip-flop are transferred to the inputs of two inverter buffer/drivers via a dual-latch, t.t.l. circuit type SN7474. Three other inverter buffer/ drivers accept the outputs of the polarity flip-flop (see Fig. 3) and drive the horizontal and vertical bars of the polarity display. The sixth inverter buffer/driver is used to provide increased power for driving some of the \(C_{p}\) inputs of the latches.
In order to understand the operation of the control logic, and hence the operation of the d.v.m., Table 3 has been drawn up. It shows the logical states of the outputs of
the relevant elements of the control logic for various steps in the operation of the d.v.m. For the moment the functions of flip-flops \(F\) and \(G\) have been ignored and the logical state of the output of NAND 1 is assumed to be a permanent logical " 1 ". For the initial conditions it will be assumed that all the logical elements are in such states as at the beginning of a measurement cycle. This is as in step 0 of the above table. No clock pulses have as yet been generated, the first decade counter and the main counter are all set to zero and the output \(\bar{C}\) is a logical " 1 ". The output \(V_{o}\) is also a logical one. The input of the integrator is thus connected to \(V_{i n}\), which is assumed to have a magnitude greater than zero but less than 2 volts. The clock is running and the first decade counter and the main counter start to count up to a total of 20,000 clock pulses (a period of 100 ms ). At some time during this period the output \(V_{o}\) will become logical " 0 " and the outputs of the various logic elements become as in step 1 of Table 3.
At the end of this period the state of the main counter is again \(0000(2000 \equiv 0000)\) but the output of \(A\) will be logical " 1 ", see step 2 . On receipt of another clock pulse, step 3, several things happen. Output \(B\) changes from " 0 " to " 1 "; \(D\) from " 1 " to " 0 ", resetting output \(A\) to " 0 "; and \(\bar{C}\) becomes 0 connecting the input of the integrator to the appropriate reference voltage. The resetting of \(A\) to " 0 " transmits a pulse to the polarity flip-flop of Fig. 3, transferring appropriate logic levels to the outputs \(P_{1}\) and \(P_{2}\). A further clock pulse, step 4, resets \(D\) to logical "l". The main counter now continues counting and nothing further happens to the logic until \(V_{o}\) again becomes logical " 1 ", indicating the end of the measurement cycle and the fact that the integrating capacitor has been discharged, see step 5 . Immediately \(V_{o}\) becomes logical " 1 ", the \(k\) input to flip-flop \(B\) becomes logical " 1 ". \(B_{k}\) is the latch pulse and a logical " \(l\) " is immediately transmitted to the latches via AND gate 4, NAND gate 3 and inverter 3. The logical states of the outputs of the main counter and flip-flop \(A\) are thus transferred to the three decoders and inverter buffer/drivers 1 and 2 , indicating the count reached by the main counter and whether there is an overload or not. Since the magnitude of \(V_{\text {in }}\) was assumed to be less than 2 volts the counter will not have reached 2000 and the output of \(A\) will be logical " 0 ", indicating no overload. On receipt of the first clock pulse after \(V_{o}\) has become logical 1 , step 6 , again a number of things happen. Output \(B\), and thus its input \(B_{k}\) (the latch pulse), changes from logical " 1 " to " 0 "; output \(E\) changes from " 1 " to " 0 ", setting output \(D\) and the output from the divide-by-two flip-flop to " 0 " and also, via NAND gate 2 , resetting all four decade counters to zero; and output \(\bar{C}\) changes from logical " 0 " to " 1 ", reconnecting the input of the integrator to \(V_{\text {in }}\). Since output \(D\) has become " 0 " output \(A\) will also be set to " 0 " if it were a logical " 1 " before. A second clock pulse, step 7, resets \(E\) from " 0 " to " 1 ", releasing the first decade counter and the main counter (allowing it to begin counting again) and flip-flop \(D\). A third clock pulse, step 8 , resets \(D\) from " 0 " to " 1 ",

Table 3. Sequence of logical states of control logic elements
\begin{tabular}{ccccccccccccccccc}
\hline Step & \(\boldsymbol{C P}\) & Count. & \(\boldsymbol{A}\) & \(\boldsymbol{V}_{o}\) & \(\boldsymbol{B}_{\boldsymbol{i}}\) & \(\boldsymbol{B}_{\boldsymbol{k}}\) & \(\boldsymbol{B}\) & \(\boldsymbol{C}_{\boldsymbol{j}}\) & \(\boldsymbol{C}_{\boldsymbol{k}}\) & \(\overline{\boldsymbol{C}}\) & \(\boldsymbol{D}_{\boldsymbol{j}}\) & \(\boldsymbol{D}_{\boldsymbol{k}}\) & \(\boldsymbol{D}\) & \(\boldsymbol{E}_{\boldsymbol{j}}\) & \(\boldsymbol{E}_{\boldsymbol{k}}\) & \(\boldsymbol{E}\) \\
\hline \hline 0 & 0 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 0 & 0 & 1 \\
1 & \(?\) & \(? ?\) & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 1 & 0 & 0 & 1 \\
2 & 0 & 2000 & 1 & 0 & 1 & 0 & 0 & 1 & 0 & 1 & 0 & 1 & 1 & 0 & 0 & 1 \\
3 & 1 & 0000 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 1 \\
4 & 2 & 0000 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 1 \\
\(\mathbf{5}\) & 0 & \(? ?\) & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 \\
6 & 1 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 0 & 0 & 1 & 0 & 0 \\
7 & 2 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 0 & 1 \\
8 & 3 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 0 & 0 & 1 \\
\hline 9 & 0 & 2000 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 1 & 0 & 1 & 1 & 0 & 0 & 1 \\
10 & 1 & 0000 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 1 \\
11 & 2 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 0 & 0 & 1 & 0 & 0 \\
12 & 3 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 0 & 1 \\
13 & 4 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 0 & 0 & 1 \\
\hline 14 & 0 & 2000 & 1 & 0 & 1 & 0 & 1 & 1 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 1 \\
15 & 0 & \(2000+\) & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 \\
16 & 1 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 0 & 0 & 1 & 0 & 0 \\
17 & 2 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 0 & 1 \\
18 & 3 & 0000 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 0 & 0 & 1 \\
\hline
\end{tabular}

In terms of logical levels: \(B_{j}=A_{i} B_{k}=B \cdot V_{0} ; C_{j}=A . \bar{V}_{0} ; C_{k}=V_{0} ; \bar{D}_{j}=D ; D_{k}=A \cdot \bar{C} ; E_{j}=E ;\) \(E_{k}=B \cdot V_{0}\)
leaving the outputs of the various logical elements exactly as at the beginning of the measurement cycle. A new measurement cycle was, in fact, begun at step 7, the moment the main counter was released after being reset.

When the input voltage is zero or its magnitude is greater than 2 volts, the operation of the control logic is slightly modified. When \(V_{\text {in }}\) is zero volts, \(V_{o}\) does not become logical " 0 " so that when the main counter reaches 2000 the conditions of the various logic elements become as in step 9. After a further clock pulse the output of \(B\) becomes logical " 1 ", so that \(B_{k}\) also immediately becomes logical " 1 " and a zero count is transferred (remembering, \(2000 \equiv 0000\) ) to the decoders. Simultaneously, the output of \(D\) becomes " 0 " and \(A\) is reset to zero, preventing an overload indication being given. At the end of three more clock pulses, steps 11, 12 and 13, the initial conditions, at the beginning of a measurement cycle, are arrived at once more.

When the magnitude of \(V_{\text {in }}\) is greater than 2 volts, the counter will become 2000 for a second time before \(V_{o}\) has become logical " 1 ". The conditions of the outputs of the logic elements when this state is reached are as shown in step 14. Having passed 2000 the


Fig. 7. Clock pulse generator, using a dual 4-input Schmitt NAND.
counter effectively starts counting again from 0000 . Nothing further happens to the logic until eventually \(V_{o}\) does become logical 1, step 15. Immediately, \(B_{k}\) becomes logical " 1 " and the state of the main counter is transferred to the decoders, etc., together with the overload information on flip-flop \(A\). Three more clock pulses, steps 16,17 and 18, reset all the logic elements to their initial conditions. Since the overload indication represents a count of 2000 , it may be added to that shown by the display to indicate a reading in excess of 2000.

The upper limits for reliable readings in excess of 2000 are determined by the precise breakdown voltages of zener diodes \(Z_{3}\) and \(Z_{4}\), and the reverse breakdown voltage of the holding transistor \(\operatorname{Tr}_{10}\) across the integrating capacitor. When \(V_{i n}\) is negative, the integrating capacitor charges positively so that \(T r_{10}\) will breakdown if the voltage across the capacitor exceeds about 5.7 volts ( \(V_{b e}=5 \mathrm{~V}\) ). The integrating capacitor charges up to a voltage level, \(V_{c}\), equivalent to
\[
(1 / R C) \int_{0}^{100 \mathrm{~ms}} V_{i n} \cdot \mathrm{~d} t
\]
i.e. \(V_{c}=V_{\text {in }} .100 \mathrm{~ms} / R C\), and since \(R C=\) \(20 \mathrm{k} \Omega \times 2.2 \mu \mathrm{~F}=44 \mathrm{~ms}, V_{c}=2.27 V_{\text {in }}\). Theoretically then, the upper limit for reliable readings in excess of 2000, when \(V_{\text {in }}\) is negative, is, \(5.7 \mathrm{~V} / 2.27=2.500\). When \(V_{\text {in }}\) is positive, the upper limit for a reliable reading in excess of 2000 is the breakdown voltage of \(Z_{3}\) plus the forward diode voltage drop across \(Z_{4}\) which in total, is about 3.300.

Having discussed the operation of the control logic of the d.v.m. in its Auto mode (by neglecting the operation of flip-flops \(F\) and \(G\) ) it is now time to consider the function of the Auto/Manual and Sample/Hold switches and flip-flops \(F\) and \(G\) in conjunction with the hold circuitry, \(I C_{7}\) and etc., of Fig. 3. As shown in Fig. 4, with the Auto/ Manual switch in the Auto position and the Sample/Hold switch as indicated, the output of flip-flops \(F\) and \(G\) are held at logical zero and hence the output of NAND gate 1
is a logical " 1 ". The operation of the control logic and the d.v.m. is therefore as described above, providing a continuous sampling of the input voltage. If now the Auto/Manual switch is switched to the Manual position, flip-fiops \(F\) and \(G\) begin to count the main counter's resetting pulses transmitted by flip-flop \(E\). After three such pulses the outputs of both \(F\) and \(G\) become logical " 1 " and hence the output of NAND gate 1 becomes logical " 0 " and that of NAND gate 2 , logical " 1 ". The main counter is thus held permanently reset and no further pulses are transmitted by flip-flop \(E\). Simultaneously, the logical " 0 " level at the output of NAND gate 1 switches the output of \(I C_{7}\) to -10 V , switching \(T r_{10} \mathrm{ON}\) and holding the voltage across the integrating capacitor to zero volts. The d.v.m. remains in this state, holding the last reading of \(V_{i n}\) on the display, until either the Auto/Manual switch is switched back to Auto, or the Sample/Hold switch (which is a simple, returning, push-button changeover switch) is depressed. If the Sample/Hold switch is depressed and released, the outputs of \(F\) and \(G\) are cleared to zero and the d.v.m. operates for a further three cycles until \(F\) and \(G\) are once more logical ones, when it again holds, displaying the latest reading of \(V_{i n}\). If, when the Auto/Manual switch is in the Auto position, the Sample/Hold switch is depressed and held down, the d.v.m. will hold its reading, after three further cycles, until the Sample/Hold switch is released. In this manner the d.v.m. can be made to continuously sample \(V_{i n}\) and hold the information for as long as is desired by depressing the Sample/Hold switch, or sample \(V_{\text {in }}\) only on demand when the Sample/Hold switch is momentarily depressed.

Having described the operation of the d.v.m., there are now bat a few points to clear up concerning the analogue circuit of Fig. 3. As mentioned earlier, the two comparators, 1 and 2 , are \(I C_{8}\) and \(I C_{9}\) respectively. The two reference voltages, -2 mV and +2 mV , applied to the two comparators are provided by the two 100 -ohm preset potentiometers and their associated circuitry. Each 1.00 -ohm pot. provides an output voltage in the range of +12 mV to -12 mV , approximately. This allows for the offset tolerance of the 709 op -amps, used for the two comparators, of about \(\pm 8 \mathrm{mV}\). The output of the integrating opamp connects to the non-inverting (positive) input of each comparator via a 10 -ohm resistor. A very small amount of positive feedback is applied to each comparator by feeding their outputs back to their respective positive inputs via a \(220 \mathrm{k} \Omega\) resistor. In practice this results in about 2 mV of hysteresis in the switching action of the two comparators. The output of comparator 2 changes from +10 V to -10 V when the output from the integrator exceeds about +3 mV , and changes back again to +10 V when the integrator output falls to about +1 mV . Comparator \(1\left(I C_{9}\right)\) operates in a similar manner about levels of -1 mV and -3 mV . This small amount of positive feedback results in a more positive switching action of the comparators without impairing their resolution. The output from the integrator is about \(2.27 V_{i n}\), thus 2 mV
hysteresis represents a resolution of about 0.88 mV , which is less than the resolution of the display ( 1 mV ).

This concludes the description of the circuitry and operation of the basic d.v.m. The rest of the article deals with the description of the a.c. rectifier circuit, a suitable input attenuator and buffer stage, power supplies and, finally, circuit board layouts and a discussion on some of the components.

\section*{References}
1. Digital Multimeter, by D. E. O'N. Waddington. Wireless World, March 1973.
2. Digital Panel Meter, by P. Bartlam. Wireless World, April 1973.

\section*{Correction}

In the article "Frequency Shifter for 'Howl' Suppression" by M. Hartley Jones (July issue, pp. 317-322) the following corrections should be made.

\section*{Appendix (p. 321)}

In Fig. 8, the point where the \(\phi_{1}\) curve crosses the frequency axis should be labelled \(f_{01}\) instead of \(f_{02}\)

The line immediately after equation (2), should read
\[
\left|\frac{v_{0}}{v_{i}}\right|=\sqrt{\frac{\alpha^{2}+\beta^{2}}{\alpha^{2}+\beta^{2}}}
\]

\section*{Components list (p. 322)}

Unfortunately there was a duplication of resistor numbers in Fig. 4 and Fig. 6. The values given in the list refer only to Fig. 4. The following additions are necessary

Components in Fig. 4
\begin{tabular}{ll}
\(R_{45}\) & \(6.8 \mathrm{k} \Omega\) \\
\(I C_{5}, I C_{6}\) & Motorola MC1495L or \\
\multicolumn{2}{c}{ Silicon General SG1495D. } \\
Components in Fig. 6 \\
\(R_{44}, R_{45}\) & \(12 \mathrm{k} \Omega\) \\
\(R_{46}\) & \(560 \Omega\) \\
\(C_{22}\) & 470 nF
\end{tabular}

In the article "F.M. Tuner Design-Two Years Later" by L. Nelson-Jones (June issue, \(\mathrm{pp} .271-275\) ), the 68 pF capacitor shown connected to the tap of \(L_{3}\) in Fig. 2 should be connected to the emitter of \(\mathrm{Tr}_{3}-\) as in the original article.

\section*{H. F. Predictions for October}

Magnetic disturbances at 27 -day intervals (one solar rotation) have been clearly evident since January of this year. Comparison with the previous sunspot cycle minimum period (as was done for the solar index in last month's notes) shows that the current disturbances are several days longer lasting than in the corresponding year of 1962. The intensity of current disturbances has dropped over the past two months, a feature also found in 1962 when, after two or three quiet rotations, the disturbances appeared consistently with each rotation for the following two years. The next most likely period of magnetic disturbance is September 23rd to October 7th.





\title{
News of the Month
}

\section*{Fifth Intelsat IV satellite}

The latest Intelsat IV communications satellite to be launched has undergone two weeks of testing before being placed in commercial service. The fifth to be put in orbit, it is positioned over the Atlantic Ocean and has a capacity to provide an average of 5,000 two-way telephone calls or 12 simultaneous colour television programmes. By the end of this year 91 aerials at 73 earth stations in 55 countries are expected to be operating with the five Intelsats.

The first two Intelsat IVs were launched in January and December 1971 over the Atlantic to provide commercial services between the U.S.A. and Europe. The third was launched over the Pacific in January 1972. The fourth was launched last June over the Indian Ocean. Each satellite is designed to have a life of seven years.

\section*{New laser-induced electrical effect discovered}

An unusual and totally unexpected electrical effect has been discovered by Robert J. von Gutfeld and Eugene E. Tynan at I.B.M's Thomas J. Watson Research Centre at Yorktown, New York.

Von Gutfeld and Tynan have found that when the surface of a thin film of a metal such as molybdenum or tungsten is irradiated with brief pulses of laser light, voltage pulses of up to 50 mV are generated in the plane of the film, for 1 kW of incident power. Such pulses can readily be detected without special amplification, and the planar direction of the voltage makes for simple attachment of electrical connections on the film surface.

Exploitation of the new effect could result in inexpensive arrays of fast photodetectors responsive over a broad optical spectrum and operable over a wide temperature range. Moreover, detectors based on the new effect would be resistant to the heat-degradation characteristic of such now-common photodetectors as silicon-based devices. Some experimental detectors based on the effect in fact show an actual increase in sensitivity with rise in temperature.

The new phenomenon was discovered during studies of heat conductivity in which a pulsed laser was used to "inject"
bursts of thermal energy into small samples of various materials under controlled and monitored conditions. A temperature gradient through the depth of the irradiated films does appear to play a central role in giving rise to the voltage, at right angles to the gradient.

At least as surprising as the voltage direction is the fact that the voltage polarity remains the same, for fixed contacts, no matter how one rotates the film in its own plane around the axis of the laser beam. The only way to reverse polarity for a sample to which measurement contacts are fixed is to shine the laser beam on the other side of the film.

In a paper in the August 15 issue of Applied Physics Letters, Dr. von Gutfeld suggests that underlying the newly found effect is an asymmetry within the films themselves - microscopic distortion caused by such factors as stress arising while the film is being deposited and/or misplacement of atoms as they stack up while the film is being formed. This type of asymmetry would be independent of rotations around the laser-beam axis and could result in the "crosswise" voltage observed, as von Gutfeld shows by an analysis of the so-called Boltzmann transport equations, which relate symmetry structure to electrical and thermal parameters. The voltage would, in fact, be a photo-induced transient thermoelectric effect.

\section*{Alphanumerics on a TV picture}

A new modular Series 204 "Display controller" from Ann Arbor Terminals, Inc., Ann Arbor, Michigan has been developed specitically for superimposing alphanumeric data on an ordinary TV picture in cable TV, closed-circuit TV, video tape recording, and annunciator applications.

The 204 has an alphanumeric display repertoire of 64 alphanumeric characters. Up to 16 lines with 32 characters per line may be displayed, and all 512 characters are stored in an internal m.o.s. dynamic shift register memory. Character size is proportional to screen size, and is typically 0.22 in on an 1 lin screen. The controller accepts either picture video or composite sync from the video source. A choice of
three output signals is provided: alphanumerics added to the picture; alphanumerics added to the composite sync; or synchronized non-composite alphanumerics that can be mixed externally with other video signals.

A switch is provided which allows the full screen of alphanumeric data to be added to the video for data display and annunciator applications, or 1,2 , or 3 lines near the bottom of the picture for titling applications.

\section*{Congress on Acoustics 1974}

Environmental acoustics will be the theme of the Eighth International Congress on Acoustics, to be held at Imperial College, London, on 23-31 July 1974. The main address at the opening plenary session, to be held in the Royal Albert Hall, will be given by R. H. Bolt (U.S.A.). This address will be on the general subject of acoustics and the environment. Other invited speakers will be continuing this theme in a series of lectures designed to review particular aspects of the subject. Offers of papers for the sessions of contributed papers are now invited, and full details on the submission of these are contained in the Second Circular, which is available from The Administrative Secretary, 8 ICA 1974, 47 Belgrave Square, London SW 1X 8QX.

\section*{Spacelab - new agreement}

Six member states of the European Space Research Organization (ESRO) have recently signed the "Arrangement between certain member states of ESRO and ESRO" for the development of the Spacelab that forms the European contribution to the American space shuttle programme. The financial contributions of the six member states concerned represent \(76 \%\) of the overall sum that Europe will devote to the Spacelab programme.

\section*{Berlin highlights}

The Berlin radio and television exhibition is without doubt the largest and most influential entertainment electronics show in Europe. Its strength lies not only in its size - 600,000 visitors, 88,000 sq.m area and 253 exhibitors - but also in its multi-level scope. It attracts top management, marketing executives, broadcasters, designers, dealers, enthusiasts and the public from all over Europe; indeed many came from Japan and the U.S.A. But despite its huge size it is relatively easy to find ones way about because of the superb organisation; moreover most of the 1973 exhibitors occupied the same places as they did in 1971.

There was much talk about a new magnetic video disc. Thought up by a private inventor earlier this year, it is presently being developed by Bogen in Berlin. Basic idea is to produce a dise that will record and playback pictures on a conventional turntable. In this idea, scanning is achieved by a record/playback head
attached to an arm that is guided by the conventional stylus-in-groove technique. Roughly half the disc has a spiral groove to guide the arm, the remaining area being treated with chromium dioxide and scanned with \(0.1 \mu \mathrm{~m}\)-gap head attached to the arm.

So far, with a rotational speed of 156 \(\mathrm{rev} / \mathrm{min}\) (chosen so that \(78 \mathrm{rev} / \mathrm{min}\) stroboscopic markings could be used) a playing time of 5 min and a bandwidth of about 2.5 MHz has been achieved (this is about four months) and work is now directed at achieving a \(12-\mathrm{min}\) playing time with a speed of \(78 \mathrm{rev} / \mathrm{min}\) and a bandwidth of 3 MHz .

Not a challenge to the Teldec video disc yet because the TED system, as it is now called, is well advanced and will be sold in Germany this coming January, the greater challenge coming from the Philips long-playing disc (see page 474, 1972), about which more next issue.

RCA's SelectaVision MagTape system was given its European première. In this \(\frac{3}{4}\)-in system, tape is retained in the cassette (called a cartridge by RCA) which means that the elaborate and no doubt expensive tape extraction mechanism of other systems is avoided.

In surround-sound systems it seems any question of standardization is being left to the market place to decide; but unfortunately not all systems are built into the hardware available. One system is not yet launched, though it was given its first public demonstration in Berlin. This is the "New Discrete" or QMX system, devised by Duane Cooper and developed by Nippon Columbia. It has the feature that software for it can be played in two quadraphonic modes, one using an inexpensive decoder and relying on the two audio channels on discs to give a very satis factory performance. For better image definition a demodulator for two carrier channels can be added, the two additional audio channels having the feature that they are narrow bandwidth and as a consequence the highest frequency on this disc is around 36 kHz rather than the 45 kHz of the CD-4 system. Both mono and stereo compatibility sound excellent.

Meanwhile, more makers on the Continent are fitting the Motorola i.c. SQ decoder in their equipment.

Further details of these developments together with a brief look at some other areas of activity will be published in the next issue.

\section*{Toshiba subsidiary in U.K.}

It's not well known that Toshiba first produced surround-sound equipment in 1964 - called "dynamic stereo". Lack of public interest meant shelving the development, but Toshiba haven't been slow in introducing other innovations, like a photoelectric cartridge, an i.c. cartridge and an electret capacitor cartridge. When the Japan market was ready for surround sound, Toshiba introduced their own matrix known as


QM and, more recently, they have developed a new stylus shape that gives similar advantages to the JVC Shibata stylus.

Now, of course, QM has given way to RM and the current Toshiba equipment is fitted with \(R M\) and \(S Q\) decoding functions.

The range of equipment now available through 600 U.K. dealers ( 50 being "hi-fi" specialists) comprises 34 products. It includes two and four-channel amplifiers and tuner-amplifiers, a matrix decoder with rear amplifiers, a stereo tuner, two openreel tape decks (one four-channel), four cassette decks (one with Philips and two with Dolby noise limiters), headphones (one with a crosstalk switch), loudspeakers and "unit audio" systems. As well, there is an 18 -in solid-state colour television receiver (Toshiba have a PAL licence) at £295, three u.h.f. black and white receivers - one 14 -in model using a 110 -degree tube - in-car equipment, and a few other items. An unusual feature on the amplifiers is a choice of turnover frequencies for bass and treble controls of 400 Hz and 1 kHz .
There is only one sad point about Toshiba's entry to the U.K. market. Data sheets we have that were printed in Japan show performance curves of equipment, but the U.K. printed ones don't.

Toshiba (UK) Ltd are at Toshiba House, Great South West Road, Feltham, Middx, telephone 01-751 1281.

\section*{Miniature Solid-state TV Camera}

An all solid-state television camera using an array of 10,000 photosensors with charge coupling, assembled on a 24 -pin dual-in-line package, has been demonstrated in the United States. Developed by the Fairchild Camera and Instrument Corporation, it measures \(3 \frac{1}{2}\) in \(\times 1 \frac{1}{2}\) in \(\times 2 \frac{1}{4}\) in, weighs six ounces and has
a power consumption of about one watt. The camera will work in conditions ranging from bright sunshine to subdued room lighting. Accessories include an optical viewfinder, tripod, monitor, pistol grip, a range of lenses and a separate unit providing battery power and radio transmission up to 100 feet. The camera is a commercial product and Fairchild say the price and availability will be announced later this year.


Solid-state miniature television camera made by Fairchild.

\section*{Briefly}

Enter the consumer. "Consumerism is not a fad," says Nathan W. Aram, a Zenith Radio Corporation vice president. "Ignoring it will not make it go away. In fact we shouldn't want it to go away, rather let's accept consumerism. Serving today's consumer is an opportunity for all of us."

\title{
Multi-channel Proportional Remote Control
}

\title{
Use of t.t.l. in low cost system giving nine channels
}

\author{
by M. F. Bessant*
}

The introduction of inexpensive servo torque units and integrated-circuit pulsewidth servo amplifiers has opened up new possibilities in the field of low-cost proportional remote control for general laboratory or industrial use. Unfortunately the associated drive circuitry available commercially is intended for model radio control, and is often built on the same printed-circuit cards as a 27 MHz transmitter and receiver. The cost-effective application of torque units and amplifiers to a system not requiring a radio link therefore depends upon the user's ability to construct suitable drive circuitry. This article outlines a remote control system offering a maximum of nine fully proportional channels, using medium scale integration t.t.l. to obtain a low component count and level of wiring complexity, at a lower cost than currently available construction kits.

\section*{Coder}

The purpose of the coder is to scan sequentially nine parallel input commands (from potentiometers for fully proportional information and switched resistors for "go/ no-go" or multi-step information) and present them to the single-line data link as a series of nine varying width pulses followed by a fixed width synchronisation pulse.

To understand the operation of the coder shown in Fig. 1, it is advisable to start on familiar ground with the collector-coupled astable multivibrator formed by transistors \(T r_{1}, T r_{2}\), and \(T r_{3}\), then assume that on the initial application of power the decode counter holds a number between 0000 and 1001 (i.e. a b.c.d. number), say 0001 . This will result in charging current being "pulled down" through channel 1 command resistor \(R_{3}\) via pin 2 of the open-collector b.c.d.-to-decimal converter, thereby allowing astable action to commence. The coder's first output pulse (taken from the collector of \(T r_{3}\) ) will be in the \(1-2 \mathrm{~ms}\) range with an exact duration determined by the setting of \(R_{3}\). The positive transition produced at the collectors of \(T r_{1}\) and \(T r_{2}\) by the termination of this pulse clocks the counter into the next state (0010) and after a 0.25 ms delay fixed by the \(C R\) time constant at the base of \(T r_{3}\), the second coder output pulse is generated (the duration of which will this time depend upon the setting of \(R_{4}\) ). All the command resistors will be sampled
sequentially in this manner until a count of 0000 is reached, when a 0.5 ms sync pulse is generated, thus "labelling" the next output pulse as a command function corresponding to channel 1 (or 0001 again).

When displayed on an oscilloscope the repeating train of nine \(1-2 \mathrm{~ms}\) varying-width pulses, with equal 0.25 ms spacing, has a distinctive "concertina" appearance (see Fig. 2(a)), with each command function being sampled approximately every 20 ms . (This coding is compatible with commercial radio-control equipment should interfacing become necessary.) In the event of a nonb.c.d. number being held in the counter at
"switch on", resistor \(R_{1}\) will enable the astable to free run at a low clock rate until one of the b.c.d.-to-decimal converter outputs goes low, preventing the system from locking up.

Fig. 1 shows channels \(1-6\) as fully proportional and channels 7-9 as "go/no-go" functions. This is only to illustrate the idea; in practice any mix of commands can be used, depending on the application.

\section*{Decoder}

The decoder accepts the serial information from the coder (via some form of data link) and by detecting the sync. pulse, passes the

nine individual commands to their respective servo amplifiers. It can be seen from Figs. 2(b) and 3 that the operation of the coder and decoder is in many ways similar due to their both being effectively clocked by opposite collectors of the same astable. Both b.c.d.-to-decimal converter outputs will therefore be almost identical (the decoder output has a 0.25 ms "offset") providing the counters are locked in step by the sync. detector clearing them both simultaneously. A change in the value of \(V R_{3}\) for example will result in a corresponding change in the duration of the negative going pulse fed to channel 3 servo amplifier via pin 4 of the decoder's b.c.d.-to-decimal converter.
Detection of the synchronization pulse is achieved by comparing the length of inverted input pulses with the output of a 0.6 ms monostable reference. Fig. 4 shows that as the minimum length of all command pulses exceeds 0.6 ms only the 0.5 ms sync. pulse presents the counter's internal "clear" NAND gate with two high inputs simultaneously, thus clearing the counter to 0000 before the arrival of the next channe! 1 command pulse. A similar combination of reference monostable and gating could be used after the decoder to detect the "go/ no-go" information pulses.

\section*{Data Link}

If the data link between the output shortcircuit protection resistor \(R_{2}\) (Fig. 1) and the decoder's input consists of more than a simple cable link (optical coupling etc.) then care must be taken not to subject the decoder t.t.l. inputs to voltages outside the decoder's supply rail limits. Transistor \(\operatorname{Tr}_{\downarrow}\) (Fig. 3) has therefore to serve the dual purpose of logical inverter and voltage clamp.
Data link bandwidth limitations present no critical problems to decoder operation for the following reasons:
(a) command pulse width information is carried on positive transitions only;
(b) these transitions are reshaped before clocking the counter by the sync. detector's Schmitt/monostable. Deterioration of the incoming pulses will not, therefore, result in reduced counter noise immunity, although excessive "pulse rounding" will eventually lead to reduced servo resolution.

Compared with the widely used technique of cascading discrete-component monostables to produce "concertina" pulse trains which are then decoded by some form of shift register (s.c.r. etc.), the approach described in this article offers many advantages. One advantage not already stated is the ability to reduce the size or power consumption of the decoder simply by substituting the standard t.t.l. shown in Fig. 3 with low power or flat pack versions where appropriate.

\section*{Servo amplifier in t.t.l.}

The system for driving six servo torque units from the m.s.i. decoder is based on torque units originally designed to provide radio control models with a reliable method of converting electrical commands into proportional mechanical movement.
A typical unit costing five pounds would


Fig. 2. Timing diagrams for (a) coder output, top, and (b) decoder output, bottom.

Fig. 3. Nine-channel decoder


Fig. 4. Sunc detector operation.

contain within its matchbox-size case a low voltage d.c. motor driving a reduction gear train, the final shaft of which connects at one end to a positional feedback potentiometer and at the other to mechanical output coupling. Backlash on this shaft would be less than \(1^{\circ}\) and stall torque approximately \(150 \mathrm{z} / \mathrm{in}\). Unloaded full drive transit time for \(300^{\circ}\) travel would be in the order of a second. These basic characteristics are compatible with low-cost, light laboratory/ industrial servo applications.

The principle of pulse proportional servo control is now well established, with the most popular types of commercially available "amplifier" (for driving the motor in the required direction to cancel errors between command and feedback pulse length) falling into the following two categories:
(a) Discrete amplifiers using push-pull motor drive that require a centre-tapped supply. Apart from the high component count (typically ten semiconductor devices plus associated passive components) these amplifiers can, in the event of power supply voltage differences, have the added disadvantage of lopsided response.
(b) Integrated circuits, custom built for radio control servo manufacturers (i.e. not available directly from semiconductor manufacturers) have the obvious size and reliability advantage over discrete counterparts, plus in some cases a bridge motor drive. They are, however, rather specialized and not easily adapted to different motor voltage, gear ratio and potentiometer resistance combinations. Both fully assembled amplifiers cost between five and six pounds

The amplifier shown in Fig. 5 is based on a t.t.l. pulse width comparator feeding a discrete bridge motor drive circuit. This combination offers a reduced component count compared with totally discrete amplifiers and improved flexibility (with comparable complexity) compared with custom i.c. amplifiers. A considerable cost saving can also be achieved if the components for all six channels are mounted on the same card (see Fig. 6). Under these conditions each t.t.l. servo amplifier will cost approximately \(£ 1\).

\section*{Circuit operation}

The position of the torque units output shaft determines the value of \(R_{T}\) which together with \(C_{T}\) and a \(2 \mathrm{k} \Omega\) resistor, form the leedback monostable's timing elements. Decoded command pulses trigger the monostable via an inverter and are compared with the resultant \(Q\) and \(\bar{Q}\) outputs. If the position requested by the command pulse differs from the output shaft's present position an error signal proportional to the difference in pulse lengths will appear at the output of either \(G_{1}\) or \(G_{2}\) open collector NAND gate depending on whether the feedback is longer or shorter in duration than the command (see Fig. 7). Provided that this error exceeds the drive amplifier's "turn-on pedestal", one side of the bridge will be turned on and the motor driven in the required direction (assuming the "sense" of the feedback is correct) to reduce the error below the turn-on level. When this is accomplished neither side of the bridge


Fig. 5. Servo amplifier using t.t./.
conducts and negligible current is drawn from the motor supply.

\section*{Expansion and deadband considerations}

After being time-division multiplexed by the coder and decoder, an individual \(1.25 \mathrm{~ms}-2.25 \mathrm{~ms}\) command will only appear at the input of its allotted servo amplifier approximately once every 20 ms . In order to sustain motor current between commands it is therefore necessary to expand the pulse length of any error produced by the comparator. As the value of the expansion components \(R_{E}, C_{D}\) and \(R_{D}\) must be equal for symmetrical servo operation only one side of the bridge will be referred to below.

The pulse expansion ratio \(N\) depends on the charge and discharge time of \(C_{E}\), together with the turn-on pedestal and is
\[
N \approx \frac{R_{E} Z_{i n}}{Z_{i n}+R_{E}} \cdot \frac{1}{R_{D}}
\]
where \(Z_{i n}\) is the drive amplifier input impedance above the pedestal. In practice \(N\) must be a compromise between servo response time and "pile up" at the higher command repetition rates (i.e. all commands set to minimum width).

Resistor \(R_{D}\) defines the minimum error pulse capable of charging \(C_{E}\) to the drive amplifier's turn on pedestal and thus cause motor current to flow. An error below this level is usually referred to as being within the "deadband". In the circuit of Fig. 5 the
width of the deadband \(t_{d}\) is
\[
t_{d} \approx \frac{C_{E} R_{D}}{4}
\]

The minimum usable deadband width is limited by the motor and gear box inertia, which may be sufficient to cause "hunting" (oscillation about the requested position). The deadband is often expressed as a percentage of command pulse modulation. For the values given we have \(t_{d}\) approximately equal to \(50 \mu \mathrm{~s}\) with 1 ms modulation; the servo is therefore said to have a \(5 \%\) deadband.

Although the expansion and deadband component values shown are not critical and can be used with most commercial units in a multi-channel system, some trade off between response time and deadband may be necessary to optimize the servo for a particular application.

\section*{Complementary bridge}

By using the complementary bridge configuration shown in Fig. 5 a wide range of motors can be driven (in either direction) from a single supply, and as any variation in this supply can only result in symmetrical changes in servo response time, the two main disadvantages associated with pushpull centre-tapped amplifiers has been eliminated. With the values shown the bridge is capable of saturation with motor stall current of up to 300 mA (typical "motor run" current is approximately


Fig. 6. Layout of components for least cost.

\section*{Books Received}

Transistor - TV Servicing Guide by Robert G. Middleton. The first chapter covers the overall subject of transistor servicing, test procedures, test equipment and basic circuit functions. Succeeding chapters describe the various picture and/or sound symptoms that may be encountered. A list of various circuit defects that could produce a particular symptom is presented, with procedures for analysing and isolating each effect. Price £2.25. Pp. 128. W. Foulsham \& Co. Ltd, Yeovil Road, Slough, SLI 4JH.

Understanding Electronic Circuits by Ian R. Sinclair explains amplifying, oscillating, switching and logic circuits, and deals extensively with the subject of integrated circuits, their merits and limitations in different applications. The level has been set for those who have some circuit wiring experience but may be uncertain of how the circuits function. Although this book is designed as a self contained work on modern circuitry, it has been written partly as a companion volume to "Understanding Elec-
tronic Components" by the same author. The two books between them offer a compact treatment covering the field of electronic components and the circuits built around them, from a practical and a theoretical point of view. Price £3.50. Pp.205. Fountain Press, Model \& Allied Publications Ltd, Book Division, Station Road, Kings Langley, Hertfordshire.

Electrical Engineering Principles and testing methods by Rhys Lewis covers fundamentals of a.c. and d.c. circuits, including network theorems, three-phase a.c. systems, transformers, d.c. machines, amplifiers, instruments and principles of testing and testing methods, the latter including the basic essentials of quality control techniques. The book is for people undertaking courses leading to technician and technologist status in electrical and electronic engineering. The latter part of the book covers the common testing methods syllabus of the City and Guilds of London Institute. Much thought has also been given to the presentation


Fig. 7. Pulse-width comparator logic.

15 mA ). Small plastic-cased transistors are quite adequate even at higher stall currents due to the very efficient saturating nature of the bridge. In order to accommodate motor voltages in excess of the SN7410N 5 -volt limit, as SN7401AN must be used, which has an open-collector rating of 15 volts. If the torque unit is capable of operating from the t.t.1. supply, decoupling between the motor and logic must be included to avoid instability.
of transforner theory. Price \(£ 4.00\). Pp. 289. Applied Science Publishers Ltd, Ripple Road, Barking, Essex.

Lightning Protection by J. L. Marshall is an examination of the phenomenon which maintains a balance in the global electrical system. It is a collation and consolidation of available information on the nature, effects and principles of protection against lightning. Its nine chapters discuss: losses resulting from lightning; the nature of lightning; magnitude of the lightning discharge; the earth as a discharge terminal for the dissipation of energy; types of protective grounding systems and methods for measuring their effectiveness; specific measures for the protection of human life; grounding communication towers and systems; protection systems for buildings and finally protection of power-transmission systems. Bibliographies are provided at the end of each chapter. Price \(£ 7.50\). Pp. 190. John Wiley \& Sons Ltd, Baffins Lane, Chichester, Sussex.

\section*{Independent Local Radio}

\title{
Preparations for sound broadcasting service opening in October
}

Commercial "Independent Local Radio" services will open in London in October with a general programme provided by Capital Radio and a special news service by the London Broadcasting Company. These will be the first two of what may eventually be as many as 60 different I.L.R. programmes in the U.K., each transmitted on both v.h.f. (Band II) and m.f. By about next spring, the London services will be joined by those for Birmingham, Manchester and Glasgow, with two more - one for Swansea, the other for Tyneside and Wearside opening in the summer of next year. Then over the next two years stations are likely to be opened in Bradford, Edinburgh, Ipswich, Liverpool, Notting ham, Plymouth, Portsmouth, Reading, Sheffield, Teesside and Wolverhampton, though not in that order. Stations are also being planned for Belfast, Blackburn, Bournemouth, Brighton, Bristol, Cardiff, Coventry, Huddersfield and Leeds. These 27 stations should provide services for just over half of the population of the U.K.

For all these services the controlling body is the I.B.A., which will build and operate all the transmitters and, in conjunction with the Post Office, provide the distribution links. The studios and studio equipment will be the concern of the programme companies, although these are required to operate within the technical characteristics set out in a detailed I.B.A. code of practice.

The planning and engineering of an entirely new series of broadcasting services is not something that happens every day - and sound radio has a life style very different from television. The problems include:
1. The search for suitable frequency allocations with the need to achieve close co-operation with the existing users of the crowded frequency spec trum.
2. The search for suitable transmitter sites (with relatively large areas needed for m.f. aerial systems) close to major cities.
3. The provision of balanced coverage on both m.f. /a.m. and v.h.f./f.m.

All these problems need to be solved without making the whole operation so costly that it would be no longer viable.
Before the Sound Broadcasting Act 1972 (now consolidated into the Indepen-
dent Broadcasting Act 1973) reached the statute book in July 1972, preliminary planning was undertaken by the Ministry of Posts and Telecommunications, and two I.B.A. engineers - J. B. Sewter and F. Wise - were temporarily seconded to M.P.T. to help in this work

The problems of planning frequencies and coverage areas were formidable. For many years the medium-wave band in Europe has appeared to be grossly overcrowded - and this is certainly the case after dark. Yet in daytime listeners have not been well served in programme choice compared with what is theoretically feasible. So what chance was there of setting up a whole new network each carrying a different programme without finally breaking the camel's back?

Then again many of the views on what constitutes an adequate m.f. signal in urban centres dates back to the days before the widespread use of steel-framed buildings and car radios. And although superficially Band II might appear less crowded large sections were then occupied by police, ambulance and similar two-way mobile communications services. Further, there were some anomalies in the B.B.C network - for example the use of Wenvoe to carry both Welsh and West regional programmes - that proved to have serious repercussions on the planning of I.L.R. stations. Again, the widening use of stereo, with its added susceptibility to adjacent channel interference, tends to emphasize the complexity of the problem. Yet another restriction arises

I.B.A. transmitter for v.h.f. coverage in London.
from the need to avoid harmonics of the 10.7 MHz i.f. channel used in v.h.f. broadcast receivers.

For m.f. only one main "U.K.-assigned" channel ( \(261 \mathrm{~m}, 1151 \mathrm{kHz}\) ) is being made available exclusively for I.L.R. stations; 1546 kHz ( 194 m ) will be shared with B.B.C. stations; other frequencies will be those internationally agreed under Article 8 of the 1948 Copenhagen Convention and Article 9 of the Radio Regulations.

Despite these various limitations, the I.B.A. engineers felt that the setting up of a new system provided an unusual opportunity to look afresh at the technical side of sound broadcasting. Much of the standard work in this field dates back to the 'thirties before the "distraction" of television.

A small team from what was then still the I.T.A. visited North America to investigate recent experience in designing and operating local radio stations in the United States and Canada. What they found made a profound impression. For example advantage was being taken by many v.h.f. stations of mixed polarization - circular or slant polarized signals - to provide a better and more homogeneous coverage for car radios and portables using telescopic v.h.f. whip aerials; again, on m.f., there was growing use of complex directional transmitting aerials using quite large numbers of mast radiators and capable of producing deep nulls in the radiation pattern to limit interference with co-channel stations. It was also noted that American practice aimed at providing relatively strong signals in the centres of towns to overcome local screening. It is not unusual to find American planning based on providing m.f. field strengths of 50 or even \(100 \mathrm{mV} / \mathrm{m}\) in city centres. One result has been that after many discussions, the contour of \(3 \mathrm{mV} / \mathrm{m}\) has now been adopted for the planning of m.f. stations for the I.L.R. system: it is no secret that higher figures were proposed.

The combination of strong signals and multiple use of the same channel may seem mutually contradictory - and this would certainly be the case were the network to depend on omni-directional transmitting aerials. By opting for threeand four-mast radiators, the I.B.A. are planning to use 1151 kHz in London, Birmingham, Manchester and Glasgow.

This again imposes important restrictions on the choice of m.f. sites. In the London area no less than 200 different sites were investigated. The need to reduce radiation towards Birmingham points to a site to the north-west of London from which signals can be directed over the main London area while minimized in the Birmingham direction.

Such sites are difficult to find - an even greater difficulty is that of obtaining planning permission for their use. Local authorities and local opinion are acutely conscious of questions of "environment" and they approach any suggestion of a site requiring the erection of a number of moderately high aerials with more than a little misgiving. It soon became


Working on the London v.h.f. transmitter.
clear to the I.B.A. that permission to use any of the possible sites in north-west London would take time, and might delay the start of the service. For that reason the decision was taken to find a temporary site where an omni-directional aerial would prove reasonably effective. This has resulted in the m.f. station at the London Transport Executive's power station at Lots Road, Chelsea, where the tall chimneys provide supports for a simple wire \(T\) aerial. The aerial has a 212 ft twin-wire top loading section with a 275 ft vertical radiating section in conjunction with a very effective earthing system which benefits from the presence of the Thames. The same aerial is used simultaneously for transmissions on 557 kHz and 719 kHz to be used by Capital Radio and London Broadcast ing Company respectively. Despite the low power - less than 500 watts e.r.p. the coverage achieved by this station is extremely good.

Clearance for the use of 557 kHz was sought in the summer of 1972 when the interference levels in the London area were shown to be low; subsequently the unauthorized ship station "Radio Veronica" off the Dutch coast moved on to this channel. Veronica causes some interference in the eastern part of the London area and more especially in areas beyond the intended coverage. I.B.A. hope to transfer these services to the permanent site at Saffron Green, near Barnet, towards the end of 1974.

For v.h.f. coverage in London a suitable site exists at the I.B.A. 405 -line television station at Croydon. This introduces to the London area the problem of v.h.f. broadcasting from different sites (the B.B.C. station is at Wrotham, Kent) but it was felt that the more central site offered significant advantages. To overcome the problem of "swamping" local
listeners a special transmitting aerial with a narrow beam in the vertical plane has been adopted (see April 1973 issue, p.175). The aerial - the first broadcast aerial in the U.K. designed for circular polarization - is of six tiers, with the result that all homes less than about a mile away "see" less than 100 watts e.r.p. compared with the 2 kW in the main lobe.

Surveys made since the start of some preliminary tests in July-August 1973 suggest that generally there is less than 6 dB difference between horizontal and vertical components of the signal, with a tendency for the vertical component to be slightly stronger towards the north. American experience suggests that although polarization is relatively unimportant in cluttered surroundings, the use of circular polarization can in clear sites give an advantage of between 6 and 12 dB where reception is on portable or car receivers. The I.B.A. intend to use circular or slant polarization at all stations where new v.h.f. aerials are installed; generally circular polarization will be used.

The v.h.f. stations are designed for pilot-tone stereo and a stereo link is being provided between the local studios and the v.h.f. transmitters - because of the local nature of this operation the problems associated with nation-wide stereo distribution do not arise. It is expected that the programme companies (other than the special news station) will make considerable use of locally originated stereo. In addition, the studios will be linked by a monophonic "music" line to the local m.f. station. There will also be a monophonic distribution link between the London news station and all stations requiring a news feed.

The engineering of I.L.R. has meant the setting up of a new local radio planning group within the I.B.A. Engineering Division and a local radio section within the existing station design and construction department; most of the other work has been achieved within the structure established for television.

What is regarded as a most important engineering pre-requisite has been the drawing up and issue of a detailed Code of Practice for the technical performance of studios and the specification of audio distortion measurements. This has provided the various programme companies with a clear idea of the technical standards expected from them. It includes, for example, sections on studio acoustics, in terms of reverberation time and ambient noise levels. It is part of the determination that I.L.R. broadcasting will be based on up-to-date engineering techniques and equipment.

Clearly companies will need some time to gain experience of their facilities time scales are such that some equipment for the London operations is likely to arrive only hours away from the start of service. But once the settling down period is over it will be fascinating to see what new head of steam is given to sound broadcasting in the United Kingdom.

\title{
Electronic Sound Synthesizer: Part 3
}

\title{
Final circuit details, interconnection of functions by patch-panel, keyboard and joystick control
}

\author{
by T. Orr \(\dagger\) *B.Sc. and D. W. Thomas \(\dagger\) Ph.D., M.I.E.R.E.
}

The final part of this series describing the construction and operation of a sound synthesizer completes the circuit functions provided with sample and hold, noise sources and the waveform generator circuitry.

\section*{Sample and hold}

It is very useful to have an analogue memory function, for use in such cases as a long fadeout where a constant control signal may be required throughout. One method of implementing this requirement is to use a sample and hold device with the following characteristics. The output should have a very small offset voltage coupled with a low output impedance; also a long storage time, so that the output voltage will only drift by a few per cent per minute; and a high accuracy over the specified input range. The sampling period is relatively short, being initiated by a positive-going pulse. Also, there is no input buffer because the output impedance of all the units of the synthesizer is low. The input voltage range is approximately -0.5 V to +6.5 V , being deliberately limited by \(D_{1}\) (Fig. 28).

The signal is stored on \(C_{3}\), a low leakage capacitor, which is connected to the input voltage by an f.e.t. \(\left(T r_{1}\right)\). This transistor is used as an analogue gate and is controlled by a monostable ( \(T_{5,6,7}\) ). During the monostable period, the gate is opened and the signal is sampled. The voltage stored on \(C_{3}\) is monitored by \(\operatorname{Tr}_{3}\), a current-driven source follower which can be preset to give a zero input/output offset voltage. Using a \(500 \Omega\) source resistor, the spread in \(V_{G S}\) may range from about -0.5 V to -5.0 V , for drain current drives from about 0.5 mA to 10 mA , respectively. The constant current source may be pre-set to lie anywhere in this range. Thus by keeping \(\operatorname{Tr}_{3}\) operating in its saturation region, and maintaining \(I_{\mathrm{D}}\) virtually constant, variations in \(V_{G S}\) can be kept very low for considerable changes in \(V_{D S}\).

Setting up procedure: set \(R_{4}\) to about \(500 \Omega\) (this is the "fine adjust" and it is preferable that \(R_{+}\)is a trimmer) and, with the input short-circuited, initiate the sampling with a positive pulse (this clears any charge on \(C_{3}\) ). Adjust \(R_{8}\) until the output voltage is as near to zero as possible and then use \(R_{\perp}\) to finely "zero" the output.

Storage time with input short-circuited is 30 minutes for \(5 \%\) droop and sampling time 14 ms .

\section*{Noise sources}

The noise sources fill two functions, firstly, a source of noise that can be filtered and modulated, and secondly, a low frequency source that can be used as a randomly fluctuating control voltage. This was achieved by constructing a white noise source and injecting the output into a spectrum shaping network and a low pass filter.

\section*{White noise source}

The major difficulty in producing a simple, reliable white noise source is the very nature of noise itself; it is non-deterministic. Several methods were available, but the simplest and cheapest seemed to be the use of the leakage current \(I_{C B O}\) of a faulty (high leakage) germanium transistor. However, this approach requires that the leaky transistor is specially selected, or even manufactured by gentle frying! A suitable device ( \(T r_{1}\) ) Fig. 29, should produce an average noise level of approximately 40 mV pk-pk, when used in the configuration shown. The white noise generator consists of three parts; the noise source \(T r_{1}\), an equalized high gain amplifier, and an output buffer. A high gain amplifier is used because the signal level from \(T r_{1}\) is relatively low, thus particular care must be taken to

Fig. 28. Circuit providing the sample and hold finctions. Resistors are \(5 \%, \frac{1}{4} W\),

isolate \(\operatorname{Tr}_{1}\) and the input of the amplifier from any power supply fluctuations. Preset \(R_{3}\) is adjusted to give a suitable output level of between 2 to 3 V pk-pk average.

\section*{Coloured noise source}

Coloured noise is produced by driving a spectrum shaping network with white noise, this network being a Baxandall tone control. Preset \(R_{17}\) is adjusted so that with both tone control pots at maximum the output shows no signs of clipping.

\section*{V.L.F. noise}

Very low frequency noise is extracted from the white noise source by two low pass filters, only one of which is available at one time, the selection being made by operating switch \(S_{1}\), Fig. 29. One of the drawbacks of this method of producing v.l.f. noise, is that very little signal remains after filtering, the amplitude rapidly diminishing with decreasing cut-off frequency.

Preset \(R_{42}\) is adjusted so that the two v.l.f. outputs have the same amplitude, of approximately \(3 \mathrm{~V} p \mathrm{pk}\)-pk average.

\section*{Waveform generator}

The waveform generator produces a control voltage that may be used to either frequency or amplitude modulate other units. The
start of the waveform is initiated by a pulse input, the output rises "exponentially" and, after a predetermined period, falls "exponentially" (Fig. 30). Three controls are provided, attack, duration and decay and the pulse may be introduced electronically or from a manual pulse source.

The circuit operation is as follows (see Fig. 31). The first section is a current driven monostable, the monostable period or duration being controlled by the current drive which is proportional to the wiper setting of \(R_{6}\). The monostable is triggered by either a positive going input pulse or from a manual pulse upon release. The square wave produced is then fed into the attack/ decay section where a capacitor is charged via the attack control \(R_{14}\) and diode \(D_{3}\). When the monostable period is over, the capacitor discharges via \(R_{12}\), the decay control, and \(D_{2}\). The potential across the
capacitor is monitored, and an attenuated and buffered output signal is produced. A choice of duration times is available ( \(C_{3}\) or \(C_{3}+C_{4}\) with \(S_{1}\) closed) and also a choice of time constants ( \(C_{6}\) or \(C_{6}+C_{7}\) with \(S_{2}\) closed).

\section*{Joystick control}

The joystick is a mechanically controlled voltage source having two degrees of freedom, and thus generating two independent control voltages, which are proportional to the stick's position. The device is essentially a position transducer (Fig. 32) with two sense pots ( \(R_{4}\) and \(R_{8}\), Fig. 33) mounted orthogonally. The range of the joystick is limited by the rectangular opening in the front panel giving approximately \(90^{\circ}\) of freedom in both the \(x\) and \(y\) directions. An extra pot can also be seen (Fig. 32) but this is used only as a spindle. The connecting

cable should be thin and flexible so as to present as little restriction as possible to the stick's movement. Also, this cable should be firmly held by two ' P ' clips, one on the joystick assembly and one on the front panel so as to stop continual wear on the soldered connections.
The circuit function is illustrated in Fig. 34. A constant potential is maintained across the control pots \(R_{4,8}\), in Fig. 33 and by the zener diodes \(D_{2,3}\). Also, the potential of these pots relative to 0 V may be shifted by presets \(R_{2.6}\). Wiper crackle is attenuated by capacitors \(C_{3,5}\) and the wiper is buffered to the output by \(T r_{2,3}\) and \(T r_{5,6}\). With the joystick in the bottom left hand corner of its range, the two outputs \(x\) and \(y\) are zeroed by adjusting \(R_{2,6}\); movement of the joystick in the \(x\) and \(y\) directions will then produce corresponding positive increases in the potential of the respective outputs.

\section*{Keyboard}

The keyboard generates a control voltage that is linearly proportional to the status of the key that is pressed. This voltage is produced for the duration of the key's depression, returning to 0 V when the key is released. If two or more keys are pressed, the highest frequency key is selected automatically. Also, when a key is pressed, a pulse is generated, this being intended to trigger the waveform generator or the sample and hold unit. However, if the



Fig. 30. Ontinus arailathe fiom the Wareform gencrator

Fig. 29. Noise source circuitry which protides white fillered or rif. monse. Resistors are \(5^{\prime \prime} \ldots \frac{1}{4} W\). capatiors \(C_{22} C_{26}\)

production of this trigger pulse is required, then care must be taken when playing the keyboard to ensure that each key is released before the next key is pressed. If this procedure is not observed, then, even though the control voltage does change correctly, no pulse will be generated. The result is the production of a signal somewhat different to that intended.

The keyboard control circuit is shown in Fig. 35. A constant potential is maintained across resistors \(R_{1}\) to \(R_{48}\), and as all these resistors are the same, they form a potential divider composed of equally spaced steps. The switches \(S_{1}\) to \(S_{49}\) are operated by the keyboard and form, with diodes \(D_{1}\) to \(D_{49}\) and resistor \(R_{51}\), a "Minof" analogue gate. Thus, whatever combination of switches are pressed, the most negative voltage is selected, this voltage appearing at the emitter of \(\mathrm{Tr}_{3}\). Note that when no switches are pressed, the emitter of \(\mathrm{Tr}_{3}\) rises to nearly \(+V_{c c}\). This voltage must be modified so that it is in a suitable form to act as a control


Fig. 32. Mechanical assembly of the joystick conirol.


Fig. 33. Circuitry assactated with the jormich contral. Resistors are \(5 \%, \frac{1}{4}\) W


Fig. 34. Ilhustration of the joystick control circuit function

Fig. 35. Keyhoard control circuit. Switches \(S_{1}\) to \(S_{49}\) are operated by the keyhoard and form with diodes \(D_{1}\) to \(D_{49}\) and resistor \(R_{51}\), a "Minof" analogue gate. Resistors are \(5 \%, \frac{1}{4} W\).
signal. It is attenuated ( \(R_{55}\) ), inverted and its d.c. level is shifted ( \(R_{59}\) ) so that the range of outputs is from 0 V to +3 V . Also, the feedback around \(I C_{1}\) is such that when no keys are pressed, and the emitter of \(T r_{3}\) rises to nearly \(+V_{c c}\), the output \(\left(V_{c}\right)\) is prevented from going negative, and stays at 0 V .

It is required that a pulse is generated at the moment when a key is pressed, but not when it is released. This would be a simple response to achieve (by detecting the transition direction of the "Minof" voltage) if it were not for the phenomenon of contact bounce. The spikes produced by the bounce can be largely suppressed \(\left(C_{3}\right)\) but there is still a possibility of generating a pulse by mistake. One method of overcoming this dilemma is to use a Schmitt trigger with a sizeable hysteresis loop, so that, as the "Minof" signal plus spikes rises or falls, it causes the Schmitt to change state only once. The direction of this change is determined by whether the input is rising or falling (i.e. whether the key is being released or pressed) and can thus be made to produce a pulse only on the falling transient.
Some applications of the keyboard are given in Fig. 36. Fig. 36(a) shows a patch

diagram of simulated piano sound. A sinusoidal signal is given a fast attack and a slow decay. Note that the control output ( \(V_{c}\) ) from the keyboard is modified by the exponential converter, so that an equally tempered scale is produced. However, if the key is prematurely released, the output promptly changes frequency. Fig. 36(b) overcomes this difficulty, by using the sample and hold circuit to store the control signal. Also, reverberation with a slow sinusoidal modulation has been added producing a pleasant effect similar to a xylophone. Fig. 36(c) shows a network for producing bell-like "clanging" noises.

\section*{Patch panel}

To provide a flexible means of programming the synthesizer, a patch panel similar to the type used in analogue computers has been included. As the input and output impedance of all the units is low, it was possible to use an unscreened system. In fact, ordinary 4 mm banana plugs and sockets were eventually chosen, this decision being greatly influenced by cost factors. This choice, however, presents a danger of damage due to misuse. If two outputs are connected together, then it is possible that some damage will eventually occur, although how long it takes is difficult to predict. Certainly, from previous experience of a similar synthesizer, no lasting damage was seen to occur when an error of this sort was made. To minimize this danger the sockets are coloured, all the inputs being yellow, the outputs being any other colour.

The synthesizers on the market appear to have overcome this difficulty, but at some cost. One method is to employ a series of horizontal and parallel conductors, one set being the inputs, the other set the outputs. Pins are then plugged in to make a connection between an input and an output, thus the danger of an "output to output" never arises. Other methods are to use switches or jack plugs instead of pins. These systems are all pre-wired and so another problem, that of the "birds' nest" of patch cords (an all too familiar sight to those who have ever used an analogue computer) has also been eliminated. However, this advantage has been gained at some expense.
The layout of the patch panel was determined on a logical basis; that is, all the oscillators on one section, the v.c.as and v.c.f. in another, the noise sources in one block etc. Also, to make connections with an external amplifier, a coax. socket was included as well as two sockets which were connected to "ground" potential, these being used as a 0 V reference point for external equipment such as voltmeters or oscilloscopes.

\section*{Power supply}

Many units of the synthesizer are sensitive to power supply fluctuations and so a stabilized supply is desirable. The circuit diagram of the supply used is given in Fig. 37. Without this suppression it is possible to trigger a response by switching on and off unconnected (except via the mains) equipment. Care should be taken in constructing the power supply to avoid introducing any high current paths that might adversely affect the circuit operation.

\section*{Appendix}

\section*{Voltage controlled filter}

Consider a bandpass filter consisting of a series \(L C R\) network. The behaviour of this system is characterized by a linear second order differential equation with constant coefficients. Using analogue techniques, it is possible to model this system, but more important it is possible to make the coefficients variable, in fact, voltage controlled.

The general equation of a linear second order system is
\[
F(t)=\ddot{x}+2 k \omega_{n} \dot{x}+\omega_{n}^{2} x
\]

Where \(\omega_{n}\) is the undamped natural frequency, \(k\) is the damping factor (note, the quality factor \(Q=1 / 2 k\) ), and \(F(t)\) is a generalized forcing function. The solution of this equation consists of two parts; the particular integral that depends on \(F(t)\), and the complementary function that depends on the solution of the right hand side only. Using the network shown in Fig. 38(a) it is possible to implement the complete solution. Different forms of \(F(t)\) can be inserted, and by varying pots 4 and 5 , the values of \(\omega_{n}{ }^{2}\) and \(2 k \omega_{n}\) can be modified. By monitoring the voltage at the output of integrator \(1(-\dot{x})\), the response of a bandpass filter, with the same coefficients, under the influence of the same forcing function \(F(t)\), is observed. (The coefficients for a series \(L C R\) circuit would be \(\omega_{n}=1 / L C\) and \(k=R / 2 C / L\) ). By monitoring \(x\), a low pass response would be seen, and \(\ddot{x}\) a high pass response. If pot 4 were an electronic


Fig. 36. Three examples of how the patch-board can be programmed for a particular sound synthesis. See text for explanation.



Pin connections for the transistors and integrated circuits used in the synthesizer's circuitry.

One method of curing both of these effects is to use two multipliers Fig. 38(b). It is easily shown that there is a linear relationship between the control voltage \(V_{c}\) and \(\omega_{n}\). Also the \(Q\) factor is invariant with resonant frequency changes (assuming multipliers 7 and 8 are matched), and the dynamic range of the filter is equal to that of one of the multipliers. It would also be possible to control the \(Q\) factor with yet another multiplier, but the use of multipliers is both expensive and introduces complications. It was for these reasons that the configuration shown in Fig. 38(c) was finally chosen. Hence, the relationship between \(V_{c}\) and \(\omega_{n}\) is "linear", the dynamic range is nearly 10 to 1 and the \(Q\) factor increases with frequency. The variation of the \(Q\) factor is not as disturbing an effect as it may appear to be, especially when it is considered qualatively.

\section*{Acknowledgements}

We wish to acknowledge the help received from Henry's Radio in the supply of certain parts, especially for the donation of the keyboard.

\section*{Capacitor ratings}

Voltage ratings of electrolytic capacitors shown in Figs. 28-38 are as follows:
Fig. \(28-C_{2} / 35 \mathrm{~V}\)
Fig. \(29-C_{1} / 25 \mathrm{~V}, C_{3} / 10 \mathrm{~V}, C_{4} / 10 \mathrm{~V}\), \(C_{7} / 10 \mathrm{~V}, C_{8} / 25 \mathrm{~V}, C_{9} / 25 \mathrm{~V}\), \(C_{10} / 16 \mathrm{~V}, C_{11} / 40 \mathrm{~V}, C_{14} / 40 \mathrm{~V}\), \(C_{17} / 16 \mathrm{~V}, C_{19} / 16 \mathrm{~V}, C_{20} / 16 \mathrm{~V}\), \(C_{27} / 10 \mathrm{~V}\).
Fig. \(31-C_{3} / 40 \mathrm{~V}, C_{4} / 16 \mathrm{~V}, C_{5} / 40 \mathrm{~V}\), \(C_{6} / 40 \mathrm{~V}, C_{7} / 25 \mathrm{~V}, C_{8} / 25 \mathrm{~V}\)
Fig. \(33-C_{1} / 25 \mathrm{~V}, C_{2} / 25 \mathrm{~V}, C_{3} / 10 \mathrm{~V}\), \(C_{4} / 25 \mathrm{~V}, C_{5} / 10 \mathrm{~V}\).
Fig. \(35-C_{1} / 25 \mathrm{~V}, C_{2} / 25 \mathrm{~V}, C_{6} / 16 \mathrm{~V}\).
Fig. \(37-C_{2} / 40 \mathrm{~V}, C_{3} / 16 \mathrm{~V}, C_{4} / 25 \mathrm{~V}\), \(C_{6} / 25 \mathrm{~V}, C_{7} / 16 \mathrm{~V}, C_{8} / 40 \mathrm{~V}\).

\title{
An approach to audio amplifier design
}

\section*{3 System design, applying the figure of merit.}

\author{
by J. R. Stuart, B.Sc. (Eng.), M.Sc., DIC, M.I.E.E.E.
}

In the second part of this series, the discussion of an approach to the design of an amplifier as part of a system led to a detailed analysis of the application of negative feedback loops. Highlighted in this analysis was the way in which the open loop characteristics of an amplifier need to be related to the closed loop operating conditions in order to achieve the correct compromise of phase, transient and steady-state distortions.

\section*{Steady-state distortions}

The transistor parameters which contribute to non-linearity have been listed in part 1 , as follows.
- The exponential form of the relationship between \(i_{b}\) and \(V_{b e}\) and of \(V_{b e}\) with temperature.
- Variations of \(h_{f e}\) and \(h_{F E}\) with collector current \(i_{c}\), with collector-emitter voltage \(V_{c e}\) (Early effect), and with temperature.

At high frequencies other effects are in variations of \(C_{b e}, C_{c b}\) and \(C_{c e}\) with chip temperature, \(V_{c e}\) and \(i_{c}\). Apart from controlling quiescent conditions, the major freedom available to the designer in defining the forward or open-loop characteristics of
*Lecson Audio Ltd
an amplifier is the choice of source and load impedance for each stage and of the amount of local feedback to be applied.

The two most useful techniques for reducing distortion introduced by device nonlinearities are local emitter feedback (in a common emitter amplifier) and the cascode configuration. Fig. 32 shows a simple common emitter amplifier with and without local feedback supplied by \(R_{e}\), and the small signal equivalent circuit for each.

We have for the case with local feedback, the trans-impedance
\[
R_{b}=\frac{V_{0}}{i_{S}}=\frac{h_{f e} R_{L} R_{S}}{h_{11}+R_{S}+R_{e}\left(h_{f e}+1\right)}
\]

Setting \(R_{e} \rightarrow 0\) gives the case of no feedback Fig. 32(a)
\[
R_{a}=\frac{V_{0}}{i_{S}}=\frac{h_{f e} R_{L} R_{S}}{R_{S}+h_{11}}
\]

By partial differentiation the sensitivity of \(R_{a}\) and \(R_{b}\) to device parameters can be shown, e.g. for change of \(h_{f e}\) for whatever reasons we have:

Case (a) no feedback
\[
\frac{\delta R_{a}}{\delta h_{f e}}=\frac{R_{L} R_{S}}{h_{11}+R_{S}}
\]


Fig. 32. Common emitter amplifier drawn (a) without feedback and (b) with feedback together with their small signal equivalents.

Case (b) with feedback
\[
\begin{aligned}
\frac{\delta R_{b}}{\delta h_{f e}}= & \frac{R_{L} R_{S}}{h_{11}+\left(h_{f e}+1\right) R_{e}+R_{S}} \\
& {\left[1-\frac{h_{f e} R_{e}}{h_{11}+\left(h_{f e}+1\right) R_{e}+R_{S}}\right] }
\end{aligned}
\]

This represents an improvement in gain stability of:
\[
\frac{\left[h_{11}+\left(h_{f e}+1\right) R_{e}+R_{S}\right]^{2}}{\left(h_{11}+R_{S}\right)\left(h_{11}+R_{e}+R_{S}\right)}
\]

Analysis will show the same improvement for many device parameters and a similar form of improvement in high frequency effects.

The cascode arrangement shown in Fig. 33 allows the common-emitter stage to be virtually freed from the Early effect and modulation of \(C_{c e}\), as the device is allowed to operate at constant \(V_{c e}\); this clearly also allows a higher bandwidth to be achieved by the stage for given source and load impedances as the Miller effect is considerably reduced.

\section*{Design of a system}

The preceding arguments in parts 1 and 2 indicate that an amplifier designed to sound very good cannot necessarily be synthesized from the basic specification:
1. Output power in excess of 40 W .
2. Power bandwidth \(20 \mathrm{~Hz}-30 \mathrm{kHz} \pm\) 1 dB .
3. Very low noise and hum, say -80 dB .
4. t.h.d. less than \(0.1 \%\) at all frequencies and power levels in the bandwidth.
5. i.m.d., however measured, less than \(0.1 \%\).
6. Low output impedance, say \(400 \mathrm{~m} \Omega\).

However it seems reasonable in the light of the preceding discussions to propose a starting point specification for very good quality as below.
1. Output power in excess of 40 W .
2. Power bandwidth \(10 \mathrm{~Hz}-30 \mathrm{kHz} \pm\) ldB.
3. Very low noise and hum, say -80 dB flat, -80 dB C.C.I.R. weighted.
4. Weighted total harmonic distortion less than \(0.1 \%\) at all frequencies and power levels; i.e. \(10 \mathrm{~Hz}-20 \mathrm{kHz}, 0-40\) watts.
5. i.m.d., however measured, less than \(0.1 \%\).
6. Low output resistance, say \(400 \mathrm{~m} \Omega\); \(10 \mathrm{~Hz}-20 \mathrm{kHz}\).
7. Open loop frequency response-any loop -3 dB at 20 kHz min.
8. Feedback factor -40 dB any loop.
9. Phase accuracy \(\pm 10^{\circ} 20 \mathrm{~Hz}-20 \mathrm{kHz}\).
10. Accurate overload characteristic inside the loops.
A typical audio amplifier system will be as shown in Fig. 34; here three major negative feedback loops are isolated. These are around the low noise input amplifier, in which equalization may be applied, the tone control stage and the power amplifier. In addition there is the volume control and a stage of filtering which need not be achieved by feedback loops.
It has been shown earlier that for any single stage to have a phase shift of \(2^{\circ}\) at 20 kHz then the minimum -3 dB closed loop bandwidth for that stage is 570 kHz ; three such stages cascaded would have a total lag of \(6^{\circ}\). It has also been demonstrated that it is not desirable to drive any audio feedback amplifier significantly above its open-loop bandwidth; therefore if the signal can be restrained to say 45 kHz in the filter stage, then the open-loop response of the two stages following the filter should be as similar as possible, thus giving a guide to the feedback factor that can be applied for a given overload margin.

The choice of 45 kHz for a passive roll-off is a compromise between the phase distortion introduced by such a filter and t.i.d. in the power amplifier. It is not in any way a magic number and may be different in every design.
At this stage the designer runs seriously short of information, in particular the extent to which phase shift can be traded off for incipient t.i.d., and this is discussed later. However, it seems reasonable to me that in view of the poor phase performance of parts of the audio chain outside direct control, e.g. the recording studio, and in view of the high apparent sensitivity of the ear to t.i.d., that it would always be preferable to err on the side of a lower passive roll-off and higher phase shift - but as a compromisenot a rule.

A recent design. A commercially available amplifying system* designed by myself is shown in block diagram form in Fig. 35.

A low-noise high overload input stage is followed by an active volume contol, filters and tone control; in each case the open-loop bandwidth and feedback factor, \(F \mathrm{~dB}\), is shown. Care has been taken to ensure that no transient distortion effects can arise with an audio signal, and the signal bandwidth of the system is constrained to 45 kHz with
a third-order Bessell roll-off which introduces a lag of \(12^{\circ}\) at 20 kHz .
It is clear from the arguments presented that, for an unconditionally stable characteristic in an amplifier which exhibits no transient distortion effects in the signal bandwidth, a low feedback factor is necessary. This is because any increase of feedback factor must be accompanied (in the general and practical situation \(\dagger\) ) by a reduction of open-loop bandwidth, \(\omega_{o L}\). The consequences of this are a rise of steadystate distortion starting below \(\omega_{O L}\) and an increased possibility of t.i.d. Therefore, in order that the amplifier should also have a weighted t.h.d. of less than \(0.1 \%\) at any frequency or power level, it was essential to achieve a low open loop distortion figure.
The final power amplifier design, which is shown in block diagram form in Fig. 36, uses a new configuration which is the subject of a British patent application.
Use of local stage feedback combined with a complementary form and output triples operating in class AB gives an openloop bandwidth of 17.5 kHz and distortion of \(0.2 \%\). The application of 32 dB of feedback reduces the weighted t.h.d. well below \(0.1 \%\) and gives an unweighted t.h.d. of \(0.005 \%\) between 100 Hz and 3 kHz .

\section*{A figure of merit}

Earlier I put forward the idea of a figure of merit which describes the quality of an audio chain or a link of that chain. This is a number derived from a weighted sum of undesirable characteristics, measured in terms of the critical parameters. This figure of merit (f.o.m.) may be time variant; that is, an amplifier may have for example a rating of 0.8 (197I) and 0.7 (1973).
It was further proposed that by using collective subjective results, any parameter could be assigned a measure of significance, and further that the starting points for each parameter would be the thresholds of perception and objection-the latter Mantel \({ }^{3}\) calls "the threshold of non-neglectability".
Successive experiments may then show improved accuracy in the choice of parameters, defining thresholds and curve fitting between the thresholds.

In this article I propose to outline a workable f.o.m. which is based on current knowledge as outtined, in the hope that its defects can be improved upon by large-scale experimental work.

The working of an f.o.m. Let us consider that the figure of merit for a chain or item in the chain be \(M\), where \(M\) is the probability that a person will not be able to detect a shortcoming in the sound. This could be restated as \(M=\) probability of non-detection of a shortcoming by a member of the population chosen at random. Therefore an ideal audio system would have an \(M=1\) and a poor system \(M=0\).
For each stage in the chain of \(n\) elements we propose \(m_{i}(i=1,2 \ldots n)\) such that the total figure of merit for the \(n\) cascaded stages is \(M_{T}=\prod_{i=1}^{n} M_{i} \alpha_{i}\), where \(x_{i}\) is a weighting factor.

Each partial f.o.m. \(M_{i}\) is composed of a weighted product of factors believed to influence the quality of the sound, such that \(M_{i}\) shows the joint probability that any one factor may be detected as a shortcoming. Therefore in a simple example, if \(M_{i}\) considers only the terms
- \(p(d)\) the probability of detection of \(d \%\) weighted t.h.d., and
- \(p(n)\) the probability of detection of \(n \mathrm{~dB}\) \(\mathrm{s} / \mathrm{n}\) ratio
then we may write
\[
M_{i}=q(d) \cdot q(n)
\]
where \(q(d)=1-p(d)\)
\(q(n)=1-p(n)\)


Fig. 33. The cascode circuit.


Fig. 34. Block diagram of a typical audio amplifier.

\footnotetext{
*The Lecson ACI + APl
tgiven output transistors
}
e.g.


Fig. 35. Black diagram of the Lecson system showing the bandwidh and feedhack factors for cach section.

Now this is clearly a simple example and does not take account of perception thresholds or interactions of parameters and masking effects. It will not be sufficient to write, for example, \(p(n)\) as the probability of detection of \(n \mathrm{~dB}\) signal to noise ratio, but possibly as a conditional probability of detection of weighted noise-say C.C.I.R. weighting--given a specific bandwidth.

So an f.o.m. which would be useful in the predictive design of audio components could be made up from tables of conditional probabilities and give a performance measure of universal use.

In this analysis I propose to use the thresholds of perception and non-neglectability \((p(x)=0\) or 1 ) for all the parameters discussed so far and to discuss interpolation between these points.

In Table 1 a list is given of these parameters, and of thresholds which seem to be reasonable in the light of current knowledge.

Frequency response is treated by considering the two roll-off points-items 1 and 2 -and determining a rough measure of \(q\left(\omega_{L}\right)\) and \(q\left(\omega_{H}\right)\) from Fig. 6 which are from results produced by Snow. Thus a response \(20 \mathrm{~Hz}-20 \mathrm{kHz}\) has a partial \(M\) of 1 while \(100 \mathrm{~Hz}-10 \mathrm{kHz}\) has a partial \(M\) of \(0.9 \times 0.95 \approx 0.86\). Phase and amplitude linearity have been considered as being logarithmically interpolated in the absence of any other information-the same method has also been used by Mantel \({ }^{3}\).

Steady state distortions are again interpolated logarithmically; this being chosen as a reasonable assumption in the absence of further knowledge. The whole basis of this experiment is to test the values and curves I have offered as a starting point.

I would suggest that at this starting point in the derivation of an f.o.m. the a priori measure of the likelihood of t.i.d.-the transient intermodulation index-be used, and the interpolation is as shown in the Table.

Signal to noise ratio is shown weighted according to the C.C.I.R. standard, and it is intended that this should include only hum and noise and not measures of crosstalk or other interfering signals.

Other parameters which have not been listed but are clearly essential when discussing elements of the chain other than amplifiers include a frequency modulation measurement to include wow, flutter and Doppler effects.

A review of the f.o.m. In the form proposed here it is possible to produce a single number which is intended to describe the subjective sound quality of a piece of equipment derived from objective measurements based on the following suppositions:
(i) it is possible to tabulate a conditional probability for the detection of any single shortcoming in terms of population.
(ii) that this probability will move in some way from 0 to 1 between the levels of perception and objection.
In order that a number may be derived, and that the behaviour of the f.o.m. may be investigated, I have used the thresholds discussed in this article, and tentatively pro-

Table 1
\begin{tabular}{|c|c|c|c|}
\hline Measurement & \[
\begin{aligned}
& p(x)= \\
& q(x)=
\end{aligned}
\] & \[
\begin{aligned}
& p(x)=1 \\
& q(x)=0
\end{aligned}
\] & Interpolation \\
\hline \multicolumn{4}{|l|}{1. Amplitude-frequency response \(20 \mathrm{~Hz} \quad 1 \mathrm{kHz}\)} \\
\hline lower -3 dB point \(\omega_{\llcorner }\) & \[
20 \mathrm{~Hz}
\] & \[
1 \mathrm{kHz}
\] & See Fig. 6 rating / 10 \\
\hline 2. Upper -3dB point \(\omega_{\mathrm{H}} \mathrm{dB}^{*}\) & \[
20 \mathrm{kHz}
\] & \[
\begin{aligned}
& 1 \mathrm{kHz} \\
& 30 \mathrm{dz}
\end{aligned}
\] & \(p(L)=0.48 \log _{10}(4 L)\) \\
\hline 3. Amplitude linearity \(\pm \angle \mathrm{dB}^{*}\) & 0.25 & 30 dB & \(p(L)=0.48 \log _{10}(4 L)\) \\
\hline 4. Phase linearity \(\pm \theta^{\circ}\). Maximum weighted th.d. or im.d.* \(d \%\) & \(5^{\circ}\)
\(0.1 \%\) & \(100^{\circ}\)
\(50 \%\) & \(p(\theta)=0.77 \log _{10}(0.20)\)
\(p(d)=0.37 \log _{10}(10 d)\) \\
\hline \begin{tabular}{l}
5. Maximum weighted t.h.d. or im.d. \(d \%\) \\
6. Transient intermodulation index
\end{tabular} & 0.1\% & 100 & \(p(d)=0.33 \log _{10}(10 t i)\) \\
\hline 7. Rise-time \(\tau\) /us & \(5 \mu \mathrm{~s}\) & 1 ms & \(\rho(\tau)=0.44 \log _{10}(0.2 \tau)\) \\
\hline 8. C.C.I.R. weighted \(s / n^{*} n\) & 70 dB & 30 dB & \(\rho(n)=(1-(n-30) / 40)\) \\
\hline 9. Cross-talk \(c\) & 60 dB & OdB & \(\rho(c)=(1-n / 60)\) \\
\hline
\end{tabular}
*In the band \(20 \mathrm{~Hz}-20 \mathrm{kHz}\) or \(\omega-\omega\) whichever is the smaller. Note \(0 \leqslant p(x) \leqslant 1\) only.

Fig. 36. Simple block diagram of the Lecson power amplifier.

posed others with an interpolation. Clearly if such an f.o.m. is shown to give accurate results then it would be of great use to designers and users of audio equipment. However, in order that a f.o.m. of this kind can evolve, very extensive listening tests should be carried out. These are probably best controlled by and published through a respected journal such as Wireless World. [We are considering this.-Ed.]

The figures tabulated in Table 1 indicate that an amplifier which equals or betters the perception threshold for every parameter would have an \(M_{i}=1\). This rating would reduce to 0.9 with a low frequency cut off of 100 Hz , or an amplitude deviation of 0.4 dB or a phase deviation of \(8^{\circ}\) etc.

\section*{Some experiments}

The author has recently carried out some listening experiments in an attempt to measure the significance of t.i.d. in high quality power amplifiers and, while the
tests are not completed, some preliminary results have been obtained which are of interest.
The approach has been to use the basic Lecson API power amplifier design and to vary only the open-loop bandwidth and feedback factor.
Three amplifiers were used:
1. The standard amplifier with an open loop bandwidth of 17.5 kHz and feedback factor of 32 dB as summarised earlier.
2. A modified version with an open-loop bandwidth of 4 kHz and feedback factor of 40 dB . This amplifier exhibited amplitude and phase responses identical to the first example, within the accuracy of the measurements \(\left(0.25 \mathrm{~dB}, 2^{\circ}\right)\), and showed t.h.d. results within \(10 \%\) weighted of the first example.
3. A modified version with an open-loop bandwidth of 17.5 kHz and a feedback factor of 6 dB . This amplifier exhibited t .h.d. of \(0.11 \% 50 \mathrm{~Hz}-3 \mathrm{kHz}\), rising to \(0.18 \%\) at

\section*{Appendix}

The three amplifiers used in these listening experiments were all of very high quality showing an f.o.m. based on the routine of Table 1 , of \(0.9,0.82\) and 0.83 , for the amplifiers \(\mathrm{A}, \mathrm{B}\) and C respectively.

Test 1. The test routine was performed using a panel of 8 listeners. Programme was derived from a very high quality disc player and monitor-standard loudspeakers employed (Spendor BC3)

Comparison A and B. On all programme material chosen, amplifier A was preferred by \(87 \%\) of the listeners. The reaction of all listeners subjectively defined a clear difference, A being preferred for greater clarity at high frequencies. On switching to \(B\) the impression was obtained of a veil being drawn over the sound, particularly with strings or percussive material.

Comparison A and C. All listeners observed audible differences; C was prefeired by \(62 \%\) on all programme material and by \(75 \%\) on folk music or percussive music. The overall impression was that C handled transient material very well but showed slight high frequency colouration, possible due to the weighted distortion

Comparison B and C. Of the total audience, \(75 \%\) preferred C on all material. Of particular interest in this test was that the two amplifiers showed subjectively different balance, C sounding to have less high frequency content than B. Also it was noted that B showed up background noise on the disc-hiss, clicks and pops-much more than C .

Test 2. A panel of 4 listeners using a high quality disc source and monitor-standard loudspeakers (Lecson HL1)

Comparison A and B. All preferred A for reasons of high frequency clarity

Comparison B and C. This test produced confusion, no direct results were applicable as preference depended upon the material used. The faults of amplifier \(B\) on transient sounds seemed to be contrasted with a slight lack of clarity on high notes with amplifier C.

Tests 3 and 4. Devised as a control test for the comparison B and C. Two panels took part, consisting of three and seven listeners respectively. Again a disc source was employed and small loudspeakers used (Spendor BCl ). In the first test C was unanimously preferred, to the second-as before, preference depended to an extent on the source material
A working hypothesis to explain reactions to amplifiers B and C could be, that subjectively the amount of t.i.d. produced by \(B\) was as significant as any high frequency t.h.d. or i.m.d. produced by C. However these listening tests are only the beginning of a serious programme of tests which will aim to establish significance over a much wider range, and so these results can only be considered to be provisional. For example no attempt has been made to establish an f.o.m. for the loudspeakers used in these tests or to calculate or measure any interactions in the reproducing systems.
\(20 \mathrm{kHz}, 35 \mathrm{~W}\) r.m.s. The distortion was such that the second harmonic was 40 dB above any other so the weighted t.h.d. was below \(0.2 \%\) at all times.

In each case the output impedance at the terminals of the amplifier was less than \(250 \mathrm{~m} \Omega 20 \mathrm{~Hz}-20 \mathrm{kHz}\), so any effect that a change of feedback factor may have had on this, was swamped by the 3 m long loudspeaker leads used.

Three experiments were conducted, two formal, one informal. In each case the amplifying equipment was arranged as Fig. 37; only two amplifiers are used in any one test and both are driven continually by the pre-amplifier. Instantaneous comparison on programme is made by switching the loudspeakers between the two power amplifiers.

In accordance with the testing procedures laid down by Percy Wilson the participants had no knowledge, until the end of the experiments, of the nature of the differences between the amplifiers (if any) nor of the kind of subjective difference (if any) to expect. At no time was it asked which of the amplifiers sounded most natural, but simply "which of two, X or Y , do you prefer?"

The results of the tests are summarized in the Appendix. It is clear that, between amplifiers which are otherwise extremely good,
despite relatively small changes to the t.i.i. performance, differences can definitely be detected by the ear as changes in the clarity and tonal balance of sound

In a future article the author intends to describe further listening and objective tests and procedures in an attempt to quantify t.i.d. in absolute terms within the f.o.m. and with respect to t.h.d.

\section*{Conclusions}

In these articles the author has attempted to study the relationships between objective tests made on amplifiers and the subjective results. Many aspects of amplifier performance have not been covered, the discussion concentrating more on distortions.

While it has been possible to outline in detail the rigorous compromises that face the designer of negative feedback amplifiers, the way in which each of the subjective effects trade-off is still not precisely known. A figure of merit calculation is given which makes an inquiring step in this direction, but it is clearly necessary that a programmed and controlled series of tests be carried out on a large scale.

\section*{Sixty Years Ago}

From time to time over the years, successive editors of Wireless World have taken issue with the Post Office on the subject of licensing, especially when it has been considered that the Postmaster-General has tried to overstep the bounds of reason by claiming a proprietorial interest in the forces of nature. A correspondent in 1913 obviously felt very much the same way. . . "We have heard lately of bedsteads and gas pipes being successfully used as substitutes for receiving aerials. Suppose I go a little further and discover that I get Paris, using only domestic appliances (such as a bedstead on an upper floor as an aerial, the wires of a piano suitably connected as a tuning coil, a nest of cake tins with buttered paper between them as a condenser, a piece of washing soda and a darning needle as a detector, and my tongue in place of the 'phones), must I obtain a licence from the Postmaster-General before I dare use such apparatus to get the time from E.L.? A few more discoveries(!) in "wireless" and we shall require to get a licence from the PostmasterGeneral before we furnish a house, and we shall have inspectors inspecting our pots and pans to see that they conform to the wireless regulations!"

\section*{Darts Game Calculator}

Apprentices at the Guided Weapons Division of the British Aircraft Corporation, Bristol, have built an experimental automatic darts game calculating system which registers and keeps scores. It comprises a special dartboard with sensing devices, and a computer-controlled display unit which acts as the scoreboard. This unit, using 120 integrated circuits, shows the running totals for the competing teams and adds up each individual score.

The dartboard is designed with each segment internally divided and connected to the display unit. Impact of a dart on each segment causes an electrical signal to be sent to the computer in the display unit. The conclusion of an individual three-dart score is signalled by the removal of the darts from the board. The system is then re-activated by the next player standing on the throwing mat, under which is concealed a proximity detector. This causes a bulb on the display unit to be lit, showing that the system is ready to accept the next score.

Any variation of the game can be fed into the display unit before the game starts, so that the starting total could be set at, say, 1001 or 301 depending on the type of game to be played. If a double is required to start the game, then a "double" light is switched on and the system ensures that electrical signals from the dartboard will not alter the setting until the first double is obtained.

The apprentices were given just 13 weeks to design, build and test the project and were allowed to spend no more than \(£ 100\) on materials.

\title{
Micropower circuits
}

\author{
by J. Carruthers, J. H. Evans, J. Kinsler and P. Williams*
}

Think small! In tune with the broadening of the frequency spectrum, so has the range of powers grown at which electronic circuits may be coerced into functioning. Within modern integrated circuits it is common to find individual transistors operating at microampere currents, with device p.ds of a volt or so. Discrete transistors can retain useful gain at currents several orders of magnitude less at room temperatures (even with silicon transistors leakage current imposes constraints on the usable current/ temperature combinations). Since the leakage currents may be markedly reduced by controlling the doping levels and depth of penetration, by attention to surface impurities and by reduction of device area, it is difficult to define this lower limit. Collector currents of 1 nA at \(20^{\circ} \mathrm{C}\) and \(1 \mu \mathrm{~A}\) at \(100^{\circ} \mathrm{C}\) are possible, though it would be foolhardy in the extreme to suggest that these could not be improved on - undoubtedly before publication some brilliant new process will appear capable of improving on these figures by a factor of ten or more!

The position is quite different in respect of minimum operating voltages. Fig. 1 shows the variation of \(I_{c}\) and \(I_{b}\) against \(V_{b e}\) for a low-leakage planar silicon transistor operated at constant \(V_{c e}\). No matter how low the current is reduced the fundamental relationship \(I_{c} \propto \exp V_{b e}\) ensures that the value of \(V_{b e}\) changes at a much slower rate. For example, a ten-fold reduction in \(I_{c}\) corresponds to a reduction of approximately 60 mV in \(V_{b e}\) at room temperatures. Since the minimum value of \(I_{c}\) is likely to be fixed by load requirements etc. no amount of juggling can reduce the minimum supply voltage below that of the corresponding \(V_{b e}\), and in most circuits the supply voltage will have to be significantly higher. Excepting special cases such as certain complementary oscillators where only one of a complementary pair needs to conduct at a time, the minimum supply voltage will be greater than 1 V and may have tö be greater than 1.5 V for op-amp type circuits.

In the above discussion no mention has been made of constraints imposed by transistor \(V_{c e}\). At low currents the value of \(V_{c e}(s a t)\) is very much less than the \(V_{b e}\) values above, though it must be observed
that \(V_{c e}\) (sat) often rises with temperature while \(V_{b e}\) always falls (typically the temperature coefficient of \(V_{b e}\) is \(-2 \mathrm{mVK}^{-1}\) ). At high current densities the bulk resistance of the semiconductor comes to dominate the junction characteristics and the minimum p.ds may well exceed a volt for both base and collector-emitter paths. Selfheating will cause the value of \(V_{b e}\) to fall somewhat but the effect should be negligible for circuits coming under the micropower heading.
The use of germanium transistors to minimize voltage requirements is well known, but the leakage currents are such that high temperatures are inconsistent with micropower operation. When using diodes, an intermediate region is provided by Schottky-barrier devices. As shown in Fig. 2, these have a p.d. appreciably less than that for a silicon diode, and can be used as bias elements as well as for rectification.

The interpretation of this term "micropower" has been made a generous one in this series because the techniques used in micropower circuits can be usefully applied in many other fields. For a given specification of load resistance and the required voltage or current swing, the minimum theoretical voltage can be determined assuming a voltage supply (if the circuit is to be supplied from a constant current then it is the minimum value of current that is defined). If this voltage is significantly greater than the minimum operating p.ds of semiconductor devices as discussed above, then standard circuit configurations may well give satisfactory results. At lower load p.ds new circuits are necessary to allow operation from correspondingly low supply voltages. As a rough guide, it is now possible to produce a.c. amplifiers, oscillators both \(R C\) and \(L C\), voltage regulators, and astable and monostable circuits that will operate from supply voltages in the region of 1 V , though with obvious limitations on output voltage swing and with reduced stability against supply/ temperature variations. Other circuits such as power amplifiers and operational amplifiers may require somewhat higher voltages but nearly all functions can be provided while operating from a single dry cell. In designing these circuits a critical parameter is the minimization of wasted voltage, and this is equally applicable to conventional


Fig. 1. Variation of \(I_{c}\) and \(I_{b}\) against \(V_{\text {be }}\) for a low leakage planar silicon transistor operated at constant \(V_{c e}\).


Fig. 2. The Schottky-barrier diode has a p.d. appreciably lower than that of a silicon diode.


Fig. 3. Replacement of a complementary pair of emitter followers (a) by common emitter stages with emitters taken to opposite sides of the supply (b).
circuits where the load-swing is to approach the available supply voltage. A good example is the replacement of a complementary pair of emitter followers Fig. 3(a) by common emitter stages with emitters taken to opposite sides of the supply Fig. 3 (b). In the former case, since the bases can only be driven with difficulty to within hundreds of millivolts of the supply, then the output suffering an additional \(V_{b e}\) loss may be up to 1 V less than the supply at both extremes, i.e. the output has a peak-peak value up to 2 V less than the supply voltage even at relatively low currents. In the circuit of Fig. 3 (b), though the drive conditions may be more difficult to meet, only the transistor \(V_{c e}\) (sat) appears in series with the load, and outputs to within a couple of hundred millivolts of the supply are feasible. It is for the same reason that the series-pass transistor is in the commonemitter mode in voltage regulators where minimum input-output differential . is important.

Two other related parameters that may be important in micropower circuits are maximum efficiency and the minimum quiescent power. As discussed in the article Power Amplifiers (June issue), these two conditions are often associated, as, for example, in class B power amplifiers. For minimum distortion it may be necessary to increase the quiescent power. Where the amplifier normally operates close to maximum output then this contribution to the power consumption is negligible. Conversely if the amplifier operates at full output only for short periods, then the mean power is strongly dependent on the quiescent power. In summary, for continuous operation at maximum outputs, saturation voltages will be the limiting factor, while the quiescent power needs most design ingenuity for large peak/mean ratios in output.

A different problem arises when efficiency at maximum output is really critical. Then the need to saturate the output transistor(s) to minimize lost voltage would bring the corresponding disadvantage that the base current becomes a large fraction of the load current. The combination of high current gain together with low saturation voltage is not an easy one, though at least high breakdown voltages are not required of the device. Special transistors called "super- \(\beta\) " devices are now used as the input stages for high input impedance operational amplifiers. These have a very thin base region, achieve gains in excess of 1,000 but have very low breakdown voltages. They are the extreme examples of another source of the trend towards low-power operation - in this case for the higher input impedance that it brings rather than for the low power itself.

The definition and control of operating current becomes difficult at low currents because of the high-value resistors needed, which are not compatible with monolithic processing in its most economic form. Circuit techniques based on the currentmirror have mitigated this problem, so


Fig. 4. At high frequencies the gainbandwidth product of a transistor is an almost linear function of quiescentcurrent.
that all the currents in an amplifier are controlled by a single low-current source. Recent micropower op-amps leave this to the choice of the user, with a single external resistor programming the operating currents of all the transistors in the i.c. At low supply voltages the p.d. across any such resistor, whether internal or external, becomes temperature-dependent and the design problems multiply.

Temperature problems are even more severe in low-voltage/low-current voltage regulators since conventional voltage reference elements cannot be used - the lowest zener diode has a breakdown voltage of approximately 2.7 V . Combinations of dissimilar diodes (e.g. Si and Ge ) can be produced that have a.voltage difference which is almost temperature compensated, while i.c. designs have exploited the properties of forward-biased silicon p-n junctions to achieve the same effect.

The one area of operation where the inherent limitations of micropower operation have not been overcome is the highfrequency region. As the quiescent current in a transistor is reduced, so the rate at which it can charge its own internal and/or external stray capacitances falls. The gain-bandwidth product is an almost linear function of quiescent current (as shown in Fig. 4) with an upper limit to this parameter short of its maximum operating current for most devices. Thus a device normally thought of as a " 100 MHz transistor", when operated at collector currents below \(1 \mu \mathrm{~A}\) may have a cut-off frequency of less then 10 kHz . Clearly it becomes of critical importance to minimize the stray and load capacitances in such applications. For micropower operations at high frequencies, transistors with the very highest quoted gain-bandwidth products should be selected -- even 1 GHz devices are not out of place provided they can sustain current gain at these low currents. Low-voltage operation brings increased problems since, for example, the collector-base diode has increasing (non-linear) capacitance as the p.d. approaches zero and eventually becomes slightly forward biased.

A major area of concern is in the digital field, where ever larger numbers of gates and other functions are being concentrated into single monolithic i.cs. These l.s.i.
(large scale integration) circuits are limited in complexity by two mechanisms - the number of external connections, and the total dissipation. Complementary m.o.s. with its extremely low standby power is the ideal logic family from this latter standpoint and is likely to dominate the market. The dissipation is significant only where high-speed operation is demanded, since then the charge/discharge of internal capacitances dissipates power. Since the choice of circuits available in this family is growing so rapidly, the user is best advised to refer to the manufacturers' data sheets, while the properties of the basic gates will be discussed in the following article.

\section*{How to get Circards}

Order a subscription by sending £9
(U.K. price; \(£ 10.50\) elsewhere) for a series of ten sets to:
Circards
J.P.C. Electrical-Electronic Press Ltd General Sales Dept.
Room 11
Dorset House
Stamford Street
London SE1 9LU
Specify which set your order should start with if not the current one. One set (normally 12 cards) costs \(£ 1\) U.K. and \(£ 1.15\) elsewhere, postage included.

Cheques should be made payable to
I.P.C. Business Press Lid.

Topics covered in Circards are
1 active filters
2 switching circuits (comparators \& Schmitts)
3 waveform generators
4 a.c. measurement
5 audio circuits (equalizers, tone control, filters)
6 constant-current circuits
7 power amplifiers (classes A, B, C, D) 8 astable circuits
9 optoelectronics: devices and applications
10 micropower circuits
Subsequent issues will cover logic gate circuits, wideband amplifiers, alarm circuits, digital counters, pulse modulators Introductory articles in Wireless World indicate availability of Circards, which are normally ready for despatch on the Ist of the month, and the Circard concept was outline in the October issue, pages 469/70.

\section*{Letters to the Editor}

The Editor does not necessarily endorse opinions expressed by his correspondents

\section*{Breakdown of c.m.o.s. devices}

I was very interested to read the article "Complementary m.o.s. Integrated Circuits" in the August issue. However, on page 396 there was the statement "To avoid failures due to destructive breakdown of the gate insulation, possible with such high impedances . . . etc".
Motorola in a recent leaflet entitled McMOS state that stress voltages can be caused by improper testing and even more likely causes are random electrostatic charges. They suggest that all leads are shorted together, antistatic clothing should be worn, personnel handling c.m.o.s. should be electrically grounded, equipment which comes in contact with finished pieces of equipment should be properly grounded - and yet they fit input protection diodes.

RCA claim to have fitted protection devices to the inputs of their devices which will have a controlled breakdown of something of the order of \(30 \mathrm{~V}-100 \mathrm{~V}\) in the reverse direction, and have given figures for the maximum voltages which can appear between pins on the i.c.
SGS also fit input protection diodes but, so far as I can ascertain, make no claims one way or the other.
This problem of breakdown due to static build-up appears to be a serious drawback to the large-scale use of c.m.o.s., or maybe I am looking at the wrong manufacturers' data sheets!

Can any of your readers enlighten me as to whether I can safely wear my nylon coat when using c.m.o.s. or do I have to chain myself and all my test gear solidly to earth?

\section*{Peter Seddon,}

Rugby,
Warwickshire.

\section*{Tuner front-end devices}

I have recehtly read in Wireless World two articles which have raised some rather large doubts in my mind as to the best device available for the reduction of cross- and inter-modulation.
In June there was Mr Nelson-Jones employing dual gate m.o.s. f.e.ts in his v.h.f. tuner which, with a sensitivity of the order of 1 microvolt, one hopes
will pick out a weak v.h.f. signal in the presence of strong local nearby frequencies.

In August P. Antoniazzi and A. Mauceri were saying the dual gate m.o.s. f.e.t. is not the answer and the bipolar transistor has a better performance, achieving a 40 dB suppression of signals with a 4 MHz separation at 500 MHz . When used as the v.h.f. part of their tuner I deduce they will claim a similar performance.

I do not have available the data sheets for the BF479; however, those for the 40673 show that careful selection of bias and a.g.c. is needed to get the best out of the m.o.s. f.e.t.

The problem I wish to pose is essentially how does one receive a 1 microvolt signal, which obviously cannot produce much a.g.c., in the presence of strong nearby v.h.f. f.m. transmissions which can easily be of the order of hundreds of millivolts.
Could these respected authors clarify this point, which is becoming a real problem with the almost saturation of v.h.f. frequency spectrum? The published literature is full of wavetraps and heli coidal filters which unfortunately are to some extent attenuators for weak signals.

Even better still would be an article from some brave soul on the relative merits of bipolars and m.o.s. f.e.ts in weak signal conditions. I am sure such an article would fill a gap in a little discussed but very real problem.
F. F. Maher,

Madrid,
Spain.

\section*{Record equalization}

In response to Mr Ewer's letter on record equalization (August issue) I would like to point out some factors which he has overlooked or perhaps does not know about.

First of all I must hold up the record companies' flag by saying that discs are recorded to the BS 1928 equalization standard and that there is no equipment "penny-pinching" when Neumann or Scully cutting lathes are in use.

The reason behind the apparent lack of extreme bass on some records is not due to the use of a modified equalization
characteristic. Despite what J. L. Linsley Hood and others say, it is due to the fact that extreme bass signals, of even moderate amplitude, can (a) damage, very easily, delicate stereo cutter heads (especially if present as a difference signal) and (b) result in groove wall breakdown, i.e. "groove-jumping". To prevent cutter head damage, fast, accurate stereo limiters are employed, but to avoid the triggering of these too often a careful watch must be kept on peak levels and difference channel levels at all frequencies, with the result that occasionally a modicum of "cut" must be applied at selected places within the audio spectrum.

There are several ways to prevent (b): (i) a deeper cut -impractical because many cheaper arms and cartridges would fail to track it; (ii) wider minimum groove spacing (before use of variable groove spacing unit) - a possible solution but it conflicts heavily with the amount of time needed per disc side for a given piece of music (Varigroove is not always the answer!); and (iii) a "roll-off" of extreme bass to allow a more nornal groove spacing.

So, Mr Ewer, don't blame the cutting engineers or their equipment: blame the totally conflicting elements of time and fidelity required from a poor little piece of black plastic and just turn up the bass!
Robert L. Arthurton,
London, N. 16.

\section*{VAT and prices}

Like your correspondent W. B. Henniker in September's letters, I have fallen foul of the "VAT addition" thing in your advertisements, in one case with further complications since by the time the VAT had been received the goods had become exhausted. At least I think they were, because I have heard nothing since the fourth of July.

Your advertisers in general were very slow to respond to VAT, and in some cases were showing advertisements in the May issue which were not clear about the tax addition.

If Wireless World depends on advertisements for a living, or if the advertisers depend on advertisements for a living, then someone will have to brighten their ideas - and their conceptions of service.
J. R. Dykes.

Winsford,
Cheshire.

\section*{Inverter for fluorescent tubes}

I was very sorry to read in the September issue of the trouble that Mr Chappell has had with the heaters of his fluorescent tube in my circuit (August issue).

I can only plead that I tested the circuit with at least three different tubes and that none of them showed any sign of the discharging from which he has suffered.

This type of tube was, of course,
designed originally for operation from the a.c. mains, and the circuits in which they are normally used arrange for the feed to the heaters for starting up to be essentially the same defined current as is used in normal running. In such a circuit discharging across the heaters is no disadvantage as the power dissipated is reduced rather than increased. Furthermore there is no reason why the heater resistance, and hence its working voltage, should be tightly controlled for such service. The value may, in fact, vary appreciably from one manufacturer to another.

Mr Chappell's suggestion of the insertion of a ten ohm resistor in series with each heater looks like a very satisfactory solution to the problem.

I am sorry that I missed this point and I hope that no other readers have suffered similar trouble.

I must apologize also for the fact that Figs 3 and 4 of my article as published had the positive power supply point omitted. This should, of course, have been at the centre-tap of the transformer just as it was in Fig. 1 (b).
K. C. Johnson,

Cheadle,
Cheshire.

\section*{Printed circuits the easy way}

My own resists are invariably photographically produced, so I cannot speak to the drawing technique which is described by J. Ferguson on page 332 of the August issue. But it ought to go on record that the only reason commercial etching of copper is done with ferric chloride (the commercial grade of which usually contains a variable proportion of free hydrochloric acid) is that this chemical is relatively inexpensive. In amateur printed circuit work there may well be other solutions which have advantages over it.

I pass over nitric acid, since it emits brown choking fumes when attacking copper. But its action is very swift, and those whose resist can stand up to its onslaught might feel it worth trying, in \(10 \%\) concentration in the open air.

My own choice would be \(10 \%\) ammonium persulphate in water. This dissolves copper very well, though perhaps a little slowly, but the solution stays fairly clear and the submerged board can be seen pretty clearly. If about \(5 \%\) by volume of strong ammonia solution is added to the persulphate, action is accelerated considerably, but the solution turns deep blue very quickly as copper is dissolved, and the effect on submerged pieces cannot be so easily seen.

Though I have not personally tried it, there would seem to be something in the electrolytic method of etching. Assuming that the unwanted copper areas of the board can be commoned, they can become an anode of the electrolytic cell, while any odd piece of scrap brass of roughly the same area can be used
as cathode. All that is now needed is a source of 6 volts 4 amps d.c. which nearly everyone must have at hand these days.

Quite a number of possible electrolytes suggest themselves, but for first trials I would suggest \(10 \%\) common salt in water. If too little current passed a board \(4 \times 4\) inches would probably need some 4 amps of current to dissolve the copper reasonably fast - the addition of a little vinegar (i.e. dilute acetic acid) should increase the conductivity to some extent. But it will not have anything like the effect that a few ml of a strong acid, such as hydrochloric acid, will produce.
P. C. Smethurst,

Bolton,
Lancs.

\section*{Electronic music}

Having been interested in the use of electronics for musical experiments, I would like to hear from any readers who have had experience in this field. I am particularly referring to the music synthesizer using voltage controlled oscillators as well as other voltage controlled functions.

The purpose of this is to collect any ideas for the design of a small synthesizer capable of fairly advanced work while still retaining its low cost.
G. Wade,

Knowle Orchard,
Churchill,
Nr. Bristol.

\section*{Projection television}

I was most interested to find reference to my old company, in conjunction with the other old stalwarts of projection television, in G. W. Tillet's Letter from America in the September issue.

It may be interesting to mention that much of the original work, culminating in the formation of White-Ibbotson Ltd, took place while I was with E.M.I. and was based on 3in. c.r.ts designed by Drs Broadway and Cairns, and using all-glass Schmidt lenses designed and made by Optical Works Ltd of Ealing. When E.M.I. decided not to proceed with the venture, I left and formed W.I. Ltd - the directors of E.M.I. giving every assistance and helpful co-operation.

Thus, we were not altogether tied to the disadvantages that Mr Tillet states applied to the Philips unit. However, I must hasten to add that neither system suffered from the dust occlusion problem mentioned if the systems were put together with proper precautions to exclude dust. In fact, the most serious problem with either the E.M.I. or the Philips c.r.t. was, in my experience, that of X-ray staining of the face glass. This could cause a \(50 \%\) light loss in a few hundred hours running.

So far as viewing "under cinema conditions" is concerned, again I must cross
swords with Mr Tillet. It is a matter of history that we had a \(4 \mathrm{ft} \times 3 \mathrm{ft}\) back projection picture running under 2.5 kW of exhibition lighting at Earls Court with a very good viewability for the passing public - and this was with a standard Philips unit. Obviously we used our lenticular screen but this provided a good horizontal angle of view and, by limiting the vertical angle to about \(15^{\circ}\) (ample for normal viewing with this size of picture), we had a minimum of projection light waste and a maximum rejection of overhead lighting interference.

It may also be interesting to record that we found a French optical company who made ordinary convex lens systems for the Philips tube with similar efficiency to the Schmidt, and at very little extra cost, which clearly avoided many of the dust exclusion/cleaning problems inherent in a basic Schmidt.

At that particular time Mullard's proved a point of view that I had always held, namely that projection would come into its own with the advent of colour. They used three primary-coloured monochrome tubes and three Philips' Schmidts. Such tubes are much less expensive to produce than the traditional panchromatic black and white tubes. The results were startingly good but orientation of the Schmidts with dichroic mirrors etc. was far from practical. Had there been colour transmission at that time, we would have used the French lenses and I think a commercially viable result would have been obtainec.
It's a great pity that we have had to leave it to the Japanese, yet again, to appreciate the advantages and go-ahead. However, I think that Sony have gone to a great deal of trouble in the wrong direction - if, of course, ours and Mullard's experiments meant anything. H. Ibbotson,

Penryn,
Cornwall.

\section*{Reflex Circuits}

I was delighted to read John ScottTaggart's letter in your September issue. First, to discover that he had not departed, nor lost interest. Second, to acknowledge his very generous final paragraph. It was all the more generous in view of the fact that I have been known to indulge in a little thinly disguised leg-pulling about such things as the celebrated ST 100 (younger readers will have to ask their elders to explain). And still more generous after I had forgotten his pioneer work on the reflex circuit. For this I must plead age, not ignorance, as I will now make clear. And I hope it will also go some way towards atoning for my sins of commission and omission as I reveal for the first time that the very book from which he quoted (Thermionic Tubes in Radio Telegraphy and Telephony) provided the foundation for my knowledge of the subject, since the degree course in electrical engineering at Edinburgh in 1921 was all "heavy". During one par-
ticular course of lectures by a notoriously ineffective expositor, while most of the inflated intake of students after the War were amusing themselves at his expense I sat in the back row, where the signal / noise ratio was fractional so did not break through my concentration, and read S-T's book from cover to cover.

As it was one of the very few "historical" books that survived my drastic pruning on moving recently to smaller premises, I really have no excuse for not refreshing my memory from it.

So far as I was concerned the ScottTaggart ubiquity to which I referred in August was mainly in print. I can recall only two sightings: one was when he came to Edinburgh about 1923 to explain and demonstrate his "Negatron"; the other was 22 years later in the Air Ministry. I am glad to acknowledge my debt to his books in the days when reliable sources of information were scarce.
M. G. Scroggie,

Bexhill,
Sussex.

\section*{"Empty" cassettes wanted}

The remedial reading department at this school uses a number of devices to encourage pupils who find difficulty in learning to read.

Our most recent innovation is a tapeloop which runs for about 18 seconds and on which are recorded about half-adozen words containing the same sounds.

The tape-loop is housed in an ordinary cassette by fitting additional rollers. Our difficulty is obtaining cassette cases. We cannot afford to purchase cassettes at normal price and then to modify them. Would any reader having empty or broken cassettes please send them to me at the school? Postage would gladly be refunded.
C. S. Smith,

The Walmer Secondary School,
Salisbury Road,
Deal,
Kent.

\section*{New names for old devices}

The headline in Electronics Weekly of August 8th reads, "Amorphous Devices Seeking a Role". Forgetting the updated terminology, these devices seem to be little more than the original coherer used in the early days of wireless to "detect" a signal. The r.f. signal fused iron filings in a tube, causing the passage of d.c. to indicate that a signal had been detected.

Newly fledged engineers should perhaps be reminded that we might never have developed an electronics industry but for Amorphous Devices.

Some readers may be interested to know that they can construct an amorphous memory device with a piece of wood, two pins and a blob of aluminium paint. The pins should be hammered into the wood about \(\frac{1}{8}\) in apart. A thin layer of paint is applied between the pins and allowed
to dry. Apply about 20 volts via a few kilohms to limit the current and the aluminium particles in the paint will link together and close a "switch" capable of carrying perhaps \(\frac{1}{3}\) amp. Apply about 6 volts without a current limiting resistor and the switch opens. It is rather like blowing a fuse. There is one difference - it can be mended again by applying the 30 volts activation supply for a moment.

I once caused great mystification amongst colleagues at the Emley Moor transmitting station with a little device able to switch a torch bulb on or off without using a switch or other familiar device. Readers wishing to repeat the experiment should beware of vibration; it will disturb the coherer link. In the event of failure please consult the paint manufacturer rather than myself.
P. J. Unwin,

Rochdale,
Lancs.
equipment; components and ancillaries.
Data networks; message switching; modems; data terminals; telemetry systems.

Mobile radio equipment and systems for land, sea and air; selective calling; radio paging; frequency allocation and channel spacing; traffic control and vehicle location systems; radio navigation receivers and systems.

Radio communications equipment and systems for fixed applications; point-to-point links; remote control systems; digital encoding systems; tropospheric scatter.

Special techniques associated with military systems; electromagnetic compatibility; routing problems; equipment parameters for future requirements; underwater communications.

Microwave communication systems; microwave antennas, ground stations and receiving systems; waveguide communication systems.

Original papers are invited from industry, government departments, the armed services, universities, etc. In the first instance, authors should submit summaries of \(200-300\) words (in English), describing the subject and scope of the paper, to: The Editor, Electronics Weekly, Dorset House, Stamford Strect, London SE 1 9LU.

Deadline for the receipt of summaries is October 26, 1973. Speakers will be notified of acceptance by December 31, 1973. Final drafts will be due by March 11, 1974.

\section*{John Gilbert retires}

After being associated with the Northern Polytechnic (now The Polytechnic of North London) for 39 years, 22 of them as head of the Department of Electronic and Communications Engineering, J. C. G. Gilbert, F.I.E.R.E., F.R.S.A. has retired. Perhaps best known to the public for his part in the television programme "Inventor's Club", his major contributions have been in the world of audio.

He was educated at Westminster City School and went on to part-time courses while, at the same time, studying music at Trinity College of Music, London, giving broadcast piano recitals between 1924 and 1928. He became interested in radio while still at school and, in 1934 joined Partridge \& Mee of Leicester on the strength of two articles on disc recording published in Wireless World, and that summer was in control of the installation of the public address system at the Regent's Park Open Air Theatre. In September of that year, Mr. Gilbert joined the "Northern Poly" as parttime lecturer on radio servicing and in 1935 became the first full-time lecturer in radio and kindred subjects. He also took up the position of technical editor of Music Trades Review. During the war years he worked under Sir Robert Renwick in the Directorate of Communications Development at M.A.P. and, on return to the Polytechnic, became senior lecturer and, in 1951, head of department - a position he held until August this year. He has been chairman of several societies, among them the British section of the Audio Engineering Society, was chairman of several B.S.I. committees and is technical consultant to the journal The Gramophone, where he continues his work. He has written several articles for Wireless World. We wish him a long and happy retirement.

\section*{Synthesizer Keyboards}

The following company has informed us that they can supply keyboards for the Wireless World sound synthesizer (see pp.485-490 this issue): Elvins Electronic Musical Instruments, 12 Brett Road, London E.8. Tel. 0.19868455.

\section*{October Meetings}

Tickets are required for some meetings: readers are advised therefore to communicate with the society concerned

\section*{LONDON}

9th. AES - "The integrated circuit in audio systems" by Jonathan A. Dell at 19.15 at the IEE, Savoy PI., WC2.

10h. IERE - Colloquium on "Electromagnetic compatibility or confusion on land, in ships and in aircraft" at 14.30 at 9 Bedford Sq., WC1.

10th. IEE - "Computers and communication convergence or conflict?" by J. R. Pollard at 17.30 at Savoy PI., WC2.

17th. IERE - "The feedback classroom" by K. Holling at 18.00 at 9 Bedford \(\mathbf{S q}\)., WCl.

18th. IEE Grads. - "The computer-oriented research laboratory by G. M. Rhodes at 18.30 at Savoy PI., WC2.

22nd. IEE/I.Mech.E. - "Sequence control of analogue computers" by Dr. G. C. Barney and Dr. D. Miller at 17.30 at Savoy PI., WC2.

23rd. IEE - "TV supply by a combination of omnidirectional 12 GHz transmission and cable distribution networks" by Dr. J. Feldmann at 17.30 at Savoy PL., WC2.

23rd. IEE Grads. - "Frequency synthesizers" by Dr. P. N. Nield at 18.30 at the Polytechnic of the South Bank, Borough Rd., SE1.

25th. IERE - AGM at 18.00 followed by presidential address by Dr. I. Maddock at 18.45 at London School of Hygiene and Tropical Medicine, Keppel St.. WC1.
26th. IEE - Discussion on "Stripline and microstrip techniques for antennas" at 17.30 at Savoy Pl., WC2.

30th. IEE/I.Mech.E - Discussion on "Remote measurement and control using Post Office data transmission facilities" at 17.30 at Savoy PI., WC2.

30th. IEE - Discussion on "Making the most of airborne frequency allocation" at 17.30 at the Royal Aeronautical Society, 4 Hamilton Pl., WI.

31st. IERE - Colloquium on "Remote control system organization" at 10.15 at Middlesex Hospital Medical School, Mortimer St., WI.

\section*{ABINGDON}

10th. IEE - "Electricity in medicine" by Dr. D. W. Hill at 19.00 at Culham Laboratory.

\section*{AYLESBURY}

16th. IEE - "Recording and reproducing quadraphonic sound" by J. C. G. Gilbert at 19.30 at Aylesbury College, Oxford Rd.

\section*{BATH}

18th. IERE/IEE - "Optical fibre communications" by F. F. Roberts at 18.00 at Lecture Room 4E3, 10, the University.

\section*{BELFAST}

IOth. IERE - Northern Ircland section AGM followed by "T.Eng. and all that" by J. T. Attridge at 18.30 at the Board Room, Ashby Institute, Queens University.

\section*{BIRMINGHAM}

3rd. IEETE - "Technician engineers and tech nicians - in the EEC, and elsewhere in the
world" by E. A. Bromfield at 19.00 at Midlands Electricity Board, Summer Lane.
29th. IEE/I.Mech.E. -Discussion on "Applications of multivariable control theory" at 18.00 at the Midlands Electricity Board, Summer Lane.

\section*{BOURNEMOUTH}

17th. IEETE - "Stereo broadcasting" by J. H. Brooks at 19.30 at The Tralee Hotel, West Hill Road, West Cliff.

\section*{BRADFORD}

18th. IEETE - "The present state of colour television" by Prof. G. N. Patchett at 19.00 at the University.
20 hh . Sept. IERE - Modern developments in hi-fi reproduction" by Dr. A. R. Bailey at 19.00 at the University.

\section*{CAMBRIDGE}

17th. IEETE - "Technician engineers and technicians - in the EEC, and elsewhere in the world" by E. A. Bromfield at 19.00 at the University Centre, Mill Lane.
25th. IERE/IEE - "Situation display - a new and unique approach to radar presentation" by F. K. H. Birnbaum at 18.30 at the University Engineering Laboratories, Trumpington St .

\section*{CARDIFF}

1Oth. IERE - "New integrated circuits for television receivers" by G. Baskerville at 18.30 at Dept. of Applied Physics, UWIST.

\section*{CHATHAM}

3rd. IERE - "Recent advances in radio navigation" by J. E. Viles at 19.00 at Medway and Maidstone College of Technology.

\section*{CHELMSFORD}

25th. Sept. IERE/IEE - "An anti-collision radar employing storage of radar pictures on tape" by J. Watt at 18.30 at the Civic Centre.

\section*{CRANWELL}

24th. IERE - "Space instrumentation" by R. Young and B. R. Kendall at 19.30 at the RAF College.

\section*{DURHAM}

17th IEETE - "TV and stage lighting" by E. Birch at 19.30 at the Lecture Theatre, Science Site, Durham University, South Road.

\section*{EVESHAM}

20th Sept. IERE - "Colour television displays the next stage" by W. W. Wright at 19.30 at the BBC Evesham Club.

\section*{FARNBOROUGH, Hants}

24th. IERE - "Multiphonic organs" by J. H. Asbery at 19.00 at Queen's Hotel.

\section*{GLOUCESTER}

18th. IERE - "Provision of communications for remote clustered visual display units" by F. B Sanders at 19.30 at the College of Technology.

\section*{IPSWICH}

17th. IERE/IEE - "Radio astronomy" by Dr. R. S. Booth at 18.30 at The Civic College.

\section*{LEICESTER}

10th. IERE - "Delta modulation systems" by R. Steele at 19.30 at the Lecture Theatre ' \(A\) ', Physics Block, the University.

\section*{LIVERPOOL}

17th. IERE - "The semiconductor story" by Dr. K. J. Dean at 19.00 at the Dept. of Electrical Engineering and Electronics, the University.

\section*{MANCHESTER}

15th. IEETE - "An engineer behind the Iron Curtain" by Prif. M. G. Say at 19.30 at Reynold Būilding, UMIST.

\section*{NORWICH}

24th. IERE/IEE - "Sonar and underwater communications" by Dr. V. G. Welsby at 19.00 at Assembly House.

\section*{PORTSMOUTH}

24th. IEETE/IEE - "The engineer in Europe" at 14.00 at the Polytechnic, Anglesea Road.

31st. IERE - "Exploring the deep oceans" by K. Haigh at 18.30 at the Polytechnic.

\section*{READING}

10th. IERE - "Digital phase lock loops" by K. Thrower and P. Atkinson at 19.30 at J. J. Thomson, Physical Laboratory, University of Reading. Whiteknights Park.

\section*{REDHILL}

24th. IEE - "Solid state radars by K. L. Fuller at 19.30 at the Mullard Research Laboratories, Cross Oak Lane, Salfords.

\section*{SOUTHAMPTON}

17th. IERE - "Charge couple devices" by J. D. E. Benyon at 18.30 at the Lanchester Theatre, University of Southampton.

\section*{SWANSEA}

24th. IERE/IEE - "Recent developments in the design of transfer function analysers" by \(W\). A. Evans at 18.15 at University College.

\section*{WOLVERHAMPTON}

16th. IERE - "The electronic control and communication network employed on the Midland links motorways by W. A. Hambrey at 19.00 at RAF Cosford, Albrighton.


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}

\title{
Realm of Microwaves
}

\section*{6. Microwave antennae}

\author{
by M. W. Hosking,* M.Sc.
}

Microwave antennae, although no different in principle to other types, are much smaller in size for a given gain. This is because of the relatively small wavelengths involved, making it possible to use aerial dimensions tens or hundreds of times the wavelength. It thus becomes more convenient and realistic to talk about an aerial aperture for transmitting or receiving power rather than an effective height for coupling to the electric field. Before discussing individual aerials it will be as well to start off with some basic definitions.

An isotropic source is an origin of energy which radiates uniformly in all directions. Placed at the centre of a sphere, the power density anywhere on the surface of the sphere would be constant. Needless to say, such a radiator is not realisable in practice, but the concept is used as a point of reference for comparing the performance of an aerial.

\section*{Gain and directivity}

There are two measures of aerial gain related by the efficiency of the aerial. The power gain \(G\) is defined by
\(G=\frac{\text { radiated power density }}{\text { power density from isotropic source }}\)
This is the figure usually quoted as it includes all the losses and imperfections associated with the antenna. Usually, gain is expressed in dB and often quoted without reference to direction, when it is then taken as applying to the peak of the main beam.

The second type of gain is the directivity \(D\) and can either be calculated or measured from the polar diagram. It is defined as
\[
D=\frac{\text { maximum power density }}{\text { average power density }}
\]
and as this does not take into account the effect of any losses it represents the ideal case. From the polar diagram of Fig. 1, the maximum of the pattern is readily seen and the average value is obtained by integrating the power density over the full \(360^{\circ}\) to give the directivity in that particular plane.

Further, as the directivity is a ratio, it is not necessary to know any absolute values; when working out the directivity from a polar plot, one reads off the peak value as a
unit of length and divides it by the mean value obtained with the aid of a planimeter. Directivity and gain are related by \(G=\eta D\), where \(\eta\) is the efficiency of the antenna. Depending on type and quality of manufacture, the efficiency of most microwave antennae lies between 50 and \(75 \%\).

From the definition of directivity, it is possible to obtain a "rule of thumb" estimate of the gain of an antenna from its \(3-\mathrm{dB}\) beamwidth. It is necessary to assume an equivalent polar diagram in which the radiated power is confined to within the solid angle formed by the product of the beamwidths in two orthogonal planes. One plane is usually taken as that parallel to the radiated electric field vector-the E-planeand the other as parallel to the magnetic field vector-the H-plane. So if this pattern has an amplitude of unity, the maximum power density is \(1 / \theta_{E} \theta_{H}, \theta_{E}\) and \(\theta_{H}\) being the half-power beamwidth in radians. The


Fig. I. General polar diagram showing the directive type of main beam that can be obtaineil in the microwave region.
Parameters such as gain, beamwidth and side-lobe level are within the control of the designei.
average power density is simply \(1 / 4 \pi\) and so the directivity is \(4 \pi / \theta_{E} \theta_{H}\), or \(41,253 / \theta_{E} \theta_{H}\) with the beamwidths in degrees. To obtain the gain an estimate must be made of the efficiency.

\section*{Sidelobes}

The great majority of polar patterns consist of a single main beam and a series of minor beams which occur at certain angular positions and the amplitudes of which decrease as the angular distance from the main beam increases. The presence of radiated power in these sidelobes is wasteful and gives a reduction in efficiency, so effort is made to reduce them. In addition, many radar systems these days have to operate in environments where they may be electrically "jammed" and the presence of high sidelobes would make them that much more susceptible. For use in low-angle tracking, high sidelobes could give rise to spurious and misleading echoes from the ground.

\section*{Antenna aperture}

It is convenient to consider microwave antennae as having an effective area which collects the incident power. Such a concept is reasonable in the case of dish types, but not so obvious for, say, a dipole. Where there is a physical aperture to measure, the effective area is less than the geometrical one by a factor which takes into account the efficiency of the aperture as a radiator. The effective area \(A_{e}\) is related to antenna gain by \(G=4 \pi A_{e} / \lambda^{2}\). The isotropic radiator, having a gain of unity, has an effective area of \(\lambda^{2} / 4 \pi\).

Impedance is important from the point of view of maximum power transfer, either transmitting or receiving, and of therefore obtaining a good match. It happens that this parameter is exceedingly complex to determine and, apart from a few very simple antennae, it has not been possible or worthwhile to compute the impedance of all types. Instead, the quite-satisfactory process of empirical matching is carried out when needed.

Having thus established some of the main parameters used to describe the performance of an antenna, this article concentrates on a widely-used type in the microwave region which radiates from an area aperture. An interesting fact about the polar pattern produced from such a radiator is that besides the gain being governed by the

physical size, both the gain and the sidelobe level are functions of the electric and magnetic field amplitudes across the aperture. One can consider the whole area to be made up of a large number of small, individual radiators, each of which is supplied with its own electromagnetic fields. By varying the field strength by different amounts at different points, it is thus possible to taper the overall aperture field in any desired way. When all radiating points have the same amplitude, the aperture is called uniformly illuminated and has its maximum gain.


Fig. 2. Main types of waveguide horn formed by flaring the waveguide walls in a controlled fashion(a) E-plane sectoral horn, (b) H-plane sectoral horn, (c) pyramidal horn, (d) conical horn.
(b)
(c)
(d)

However, this distribution also gives the highest sidelobe level. By illuminating the aperture in various non-uniform ways and accepting a slight decrease in gain, together with widening of the main beam, the sidelobes may be reduced. A widely-used distribution, for instance, follows a cosine law, the electric field being zero at the edges of the aperture and rising to a maximum at the centre. In this case, a reduction in sidelobes of about 10 dB together with a decrease in gain of IdB would be obtained compared with a uniform illumination on a rectangular aperture.

Fig. 3. For a fixed length of horn, directivity increases as the aperture widens and passes through a maximum value. Composite plots are shown here of the directivities of (a) E-plane and (b) H-plane sectoral horns having various lengths. Dimensions for an optimum horn can easily be seen.

A practical form of antenna, often used as the feed for a larger reflector-type, is simply open-ended rectangular or circular waveguide. In this case, the aperture distribution is governed by the field pattern within the waveguide, which in turn depends on the particular mode it is supporting. Usually the dominant mode is the one of interest and the Table (p. 502) lists the main beam dimensions for rectangular and circular guide when just the dominant mode is present. Both gain and beamwidth of the waveguide radiator are proportional to the area of the aperture, so that to obtain a narrow beam or high gain, a large waveguide would be needed.

As shown in part 2 of this series (March) once a certain, fairly limited, range of waveguide dimensions had been exceeded, then higher-order modes can propagate. So that using oversized waveguide to obtain a larger effective area is not recommended; any higher modes appearing would serve only to waste power and to distort the radiated pattern. Instead, the technique used is to flare out the waveguide dimensions in a controlled way to form a waveguide horn. If only two sides of a rectangular guide are flared, this is called either an \(E\) or H -plane sectoral horn as in Fig. 2; while flaring in other directions as well produces pyramidal and conical horns.

Higher-order modes can be generated at a discontinuity in a transmission line, although in the case of a guide below cut-off, they are rapidly attenuated. The horn flare angle must be chosen so that any highermodes generated at the throat of the horn are suppressed by the time they reach horn dimensions wherein they can propagate. Within this constraint, the gain of a sectoral horn increases with increasing aperture area.

If the axial length \(L\) of the horn is kept constant, together with the height \(b\) (in the case of the E-plane horn), and the aperture then widened, the gain steadily increases, passes through a maximum and then decreases. Horns with dimensions corresponding to this maximum are called optimum horns and Fig. 3(a) shows the variation in directivity of the E-plane sectoral horn for various axial lengths. Similar results obtain for the H -plane horn with constant \(L\) and \(a\), and are shown in Fig. 3(b).

By combining the E and H-plane flares, the pyramidal horn is produced and, because the radiating mechanism is well-understood, enabling the gain to be accurately calculated, this type of horn is often used as a gain standard. Theoretically, the directivity of the pyramidal horn can be obtained from the sectoral horn directivities, \(D_{E}\) and \(D_{H}\) and is \(D_{E} D_{H} \pi \lambda^{2} / 32 a b\). Optimum dimensions for this type of horn are summarized in Fig. 4(a). For example, a horn with a \(25-\mathrm{dB}\) directivity and of optimum proportions would have a length of 20 wavelengths and an aperture of 8.3 by 6.7 wavelengths.

Finally, circular waveguide can be used to produce the conical horn, the optimum dimensions versus directivity of which are plotted in Fig. 4(b). The dimensions of pyramidal and conical horns are very similar for a particular gain so, in that respect, there is no advantage of one over the other. The conical shape is more suited to a circularly-polarized antenna, but the useful waveguide bandwidth is lower in circular than in rectangular guide.

\section*{Reflector antenna}

A widely-used class of antenna fully exploits the advantages of the small microwave wavelengths to produce a highly-directive narrow beam from a conveniently-sized aperture. This is the reflector antenna and consists of a small radiator called the primary source or feed, which is used to illuminate a large reflecting dish which reflects radiation into space in the form of a concentrated beam. The feed can be of any convenient design, but the pyramidal or conical horn is widely used. For the main dish a paraboloidal contour finds most application, the geometry of the system being shown in Fig. 5 together with some important relationships.

The parabola has two important properties which account for its wide usage. Firstly, with the feed placed at the focus of the parabola, reflected rays are concentrated within a beam parallel to the axis. Conversely, when used as a receiver, the incident radiation on the main dish will be focused to the one point on the axis. Secondly, the path lengths of rays from the focus to the reflector and out into the distance are the same. This means that, for a small feed and large dish, the feed appears as a point source with a spherical wavefront which is converted by a parabola into a plane wavefront having a uniform phase. Because the reflecting dish is usually several tens of wavelengths in diameter high gains can be achieved, 30 to 45 dB being typical, resulting in beamwidths of less than \(2^{\circ}\) and making the system eminently suitable for targettracking radar or as part of a low signalstrength receiver.

When designing a parabolic reflector, one of the first parameters to settle is the ratio of focal length to dish diameter \((F / D)\). This is determined by both mechanical and electrical considerations, a small ratio meaning a deep dish and a large ratio meaning a very shallow one, both have their problems when it comes to mounting both dish and feed. In addition, the small diameter, large focal length dish requires a larger feed to produce the narrower illu-


Fig. 4. In similar fashion to the sectoral horns, the pyramidal and conical versions also have optimum aperture dimensions for a fixed length. Directivities are plotted for (a) pyramidal horn and (b) conical horn.

Fig. 6. For a quick estimate of feed-horn performance, this graph presents the halfangle \(\psi\) as a function of \(F / D\) ratio.
minating beam, thereby introducing more aperture blocking, while the large dish makes feed design difficult in obtaining a uniform phase. Generally, parabolas have an \(F / D\) ratio of between 0.3 and 0.5 .

Another important factor in this type of antenna and one which affects the overall efficiency is the variation in field amplitude across the aperture of the reflector. For maximum gain, a uniformly illuminated aperture is required, and to obtain that with this system, a paraboloid with large \(F / D\) ratio and a very wide-beam feed would be needed. With such a feed, a lot of the radiated power would spill over the edge of the reflector and be lost, defeating the object. On the other hand, the dish diameter could be increased, giving a small \(F / D\) ratio and intercepting more of the feed radiation. Efficiency would then fall because of the departure from a non-uniform field distribution. There would thus seem to be a compromise situation between these two effects


Fig. 5. Paruboloid reflector can produce high directivity and a narrow, pencil beam and allows side-lobe level control by proper design of the feed. Because all path-lengths from the feed to the far field are equal, the paraboloid converts the initially spherical wavefront into a plane-wavefront, collimated beam.

and it is found in practice that optimum efficiency occurs for a reflector illumination such that the power level at the edge is 10 to 12 dB below that of the centre. In this case, the level of the first side-lobes below the main beam is about 23 dB .

Often, maximum gain is not all-important and a lower side-lobe level is desired, so for a sacrifice of 1 to 2 dB of gain, decreasing the edge illumination to 20 dB gives a side lobe level of about -26 dB . Clearly then, the radiation pattern of the feed is of utmost importance in controlling the efficiency of the system.

When assessing the design of a paraboloidal reflector, some useful relationships can be presented in graphical form. Equation 2 of Fig. 5, for instance, is plotted in Fig. 6 to give the subtended half-angle of the feed on the dish. Having chosen \(F / D\), the half-angle can be used to determine the radiation pattern of the feed required to produce a particular illumination taper.

First, a correction factor must be applied Referring to Fig. 5, the path length of a ray, \(r\), increases as its axial angle \(\theta\) increases. So as the power level of a spherical wavefront varies as the inverse square of path length, there is already some aperture taper across the paraboloid. The amount of attenuation is \(20 \log _{10}(r / F)\) which from equation I can be written as \(20 \log _{10} \sec ^{2}(\theta / 2)\) and is plotted in Fig. 7.

Suppose we had a paraboloid with an \(F / D\) of 0.35 and we require an edge taper level of -15 dB . Fig. 6 shows that the sub tended half-angle is about \(71^{\circ}\), resulting in an inverse-square attenuation of 3.6 dB This means that 3.6 dB of the required 15 dB is inherent in the system and, therefore, the feed has to have a radiation pattern which is only 11.4 dB down on its peak at an angle of \(71^{\circ}\)

By taking into account the E and H-plane radiation characteristics of the feed, together with the dish illumination required, it is possible to define an overall directivity for the reflector system. The useful expression emerging is \(D=D_{E}+D_{H}+10 \log _{10}\) \(A / \lambda^{2}(\mathrm{~dB})\) where \(A\) is the area of the reflector, and equals \(\pi D^{2} / 4\) for a circular parabola. Directivity factors \(D_{E}\) and \(D_{H}\) take into account the illumination taper and the type of feed. A popular feed is the pyramidal horn and for this case the directivity factors are plotted in Fig. 8. Thus, if the reflector of the previous example was required to have a directivity of 40 dB at a frequency of \(20,000 \mathrm{MHz}, D_{E}=4.8 \mathrm{~dB}\) and \(D_{H}=4.5\) dB. So, \(10 \log _{10} \pi D^{2} / 4 \lambda^{2}=30.7 \mathrm{~dB}\) and the reflector diameter would need to be 58 cm . In practice, there are more losses associated with the system and typical efficiencies lie between 55 and \(65 \%\)

Finally, from the relationship given earlier between beamwidth and directivity, a rough estimate of the \(3-\mathrm{dB}\) beamwidth is \(70 \lambda / D\) degrees. In the above example, the beamwidth will be just under \(2^{\circ}\). This serves to demonstrate the highly directive type of beam which can be produced by this type of moderately-sized antenna at microwave frequencies.

One problem that exists with this type of antenna arises from the physical presence of the feed in the aperture of the main reflector. Firstly, the feed and its support introduce aperture blocking or shadowing which has the effect of reducing the gain and degrading the sidelobe level. As the transverse feed dimensions do not change markedly with dish diameter, this effect obviously gets worse for smaller reflectors. Secondly, there is interaction between feed and dish, in that energy reflected from near the axis of the paraboloid enters the feed aperture and interacts with the primary radiation. The effect is similar to a condition of mismatch and impairs the radiation efficiency of the feed. To counteract this, various matching devices can be incorporated in the feed, or, as shown in Fig. 9(a) a plate can be placed at the apex of the paraboloid and its size and position adjusted until the reflected signal is equal in amplitude but opposite in phase to signals arriving from other parts around the apex. Under these conditions, cancellation of the mismatch occurs.

Fig. 9(b) illustrates another technique: that of offsetting the feed. The feed is still placed at the focus of the paraboloid, but is inclined to illuminate an off-centre section and the remainder of the reflector can be removed. There is no longer any aperture


Fig. 7. Natural \(1 / r^{2}\) variation in radiated energy contributes towards the aperture taper and must be taken into account when designing the feed.


Fig. 8. Directivity of the complete parabolic aerial can be estimated with relation to the feed parameters. Plotted here are \(E\) - and \(H\)-plane directivity factors for a pyramidal horn.


Fig. 9. Aperture blocking by the feed reduces the gain and degrades the side-lobe level of the paraboloid antenna and several methods exist for reducing the effect. Two are shown (a) raising the apex of the dish and (b) offsetting the feed.
blocking, nor any feed/reflector mismatch effects. The main advantage of the offset feed is that it gives several dB improvement in sidelobe level. The aperture field distribution is no longer symmetrical and at offset angles greater than about \(15^{\circ}\), trans-versely-polarized field components and an effective broadening of the radiation pattern start to significantly reduce the gain.

\section*{Cassegrain antenna}

A widely-used variation of the simple paraboloid reflector is the Cassegrain system, which operates on the same principles as the optical versions in telescopes. Fig. 10 shows the general layout, together with some important geometrical relationships. An extra element has been added in the form of a hyperboloid sub-reflector and the position of the primary feed has been changed. The hyperboloid has two focal points, one real and one virtual, which are made to coincide with the feed position in one case and the focus of the paraboloidal main reflector in the other. The effect of this arrangement can be seen from the raytracing of Fig. 10(a) where an image of the real feed is produced at the virtual focus. Thus, as far as the paraboloid is concerned, it is being illuminated from this point and we have the system already described. Because of the magnifying properties of an hyperbola the image feed has a smaller effective area than the real one, but a correspondingly broader beamwidth and this can be used to advantage in cases where the feed is bulky.
One of the raisons d'etre of the Cassegrain system is that it has an effective focal length which is larger than the focal length of the paraboloid by a factor equal to the magnification. A paraboloid with a particular \(F / D\) ratio can be made to have the same effect as one with a larger ratio by using the Cassegrain system. As regards tapering the aperture illumination of the main dish, this is done in similar fashion to that already described. The difference in the Cassegrain system is that the focal length of the paraboloid is now the effective focal length. For simplicity, only the true Cassegrain parabola/hyperbola system has been mentioned, but to obtain various combinations of beam shape and aperture-blocking, the subreflector can be varied from convex to flat to concave and can also be elliptical. Similarly with the main reflector
Aperture blocking itself is more serious with the Cassegrain type of antenna because the sub-reflector tends to be larger than a simple feed. Several methods exist for minimizing this, one of which depends on an optimum choice of dimensions for both feed and sub-reflector. Fig. 10(a) shows that the sub-reflector diameter may be reduced either by bringing it closer to the feed, or by making the feed itself more directive. After a certain point the stage is reached where the shadowing caused by the feed on the paraboloid is greater than the blocking due to the sub-reflector. There thus exists an optimum when the shadows projected by the sub-reflector and feed are equal in area.
A second technique, the principles of which will be covered in a later article, makes use of the fact that it is not necessary


Fig. 10. Cassegrain antenna uses an additional element in the form of an hyperbolic sub-reflector and has an effective focal length which is longer than the focal length of the parabola. An image of the feed is formed at the virtual focus and illuminating radiation appears as coming from this point.
to have a solid metal dish to reflect energy. A properly designed grating of wires will do the job just as well, reflecting radiation whose electric field vector is parallel to the wires. For an E-vector perpendicular to the wires, the grating appears transparent and the radiation can pass through undisturbed. Also, by using a wire grating, it is possible to construct a reflector which will rotate the polarization of the incident radiation. So the sub-reflector could be made of a grid of horizontal wires and the main parabolic dish could incorporate a polarizationtwisting arrangement. Horizontally-polarized radiation from the feed would be reflected from the sub-reflector onto the main dish and would emerge as a verticallypolarized field, to which the sub-reflector is transparent. Thus-no aperture blockage by the sub-reflector. It is evident that this system can only be used with singlypolarized antennae.

Besides the ability to tolerate a large feed, the Cassegrain antenna has several other advantages over the simple parabolic reflector. Having the feed tucked away at the rear of the dish eliminates the relatively long waveguide run and the associated losses. Although perhaps only a fraction of a dB , this is important to low-noise receivers such as might be used in radio-telescope and communication systems. Because of the positioning of the feed, there is also less noise introduced into the Cassegrain system by spill-over radiation being reflected from the ground.

\title{
Literature Received
}

\section*{For further information on any item include the \(W W\) number on the reader reply card}

\section*{ACTIVE DEVICES}
"Laser trimming techniques for thick film resistors" is a 6 -page publication describing the advantages and use of laser trimming techniques. DuPont de Nemours International S.A., Post Office Box CH-1211, Geneva 24, Switzerland ....... WW 401

\section*{Passive devices}
"Precision Self-latching Electrical Connectors" is a wall chart containing details of the sizes. make-up, and electrical characteristics of standard Lemo connectors from size 00 to 06 . Lemo (U.K.) Ltd, Worthing House, 6 South Street. Worthing, Sussex BN 11 3AE

\section*{EQUIPMENT}

We have received a data sheet describing the series RO200C Paraliel Display Controllers which provide alphanumeric data display from a computer data bus. Output is composite video, compatible with E.I.A. standard 525 -line video monitors. Ann Arbor Terminals Inc., 6107 Jackson Road, Ann Arbor, Michigan 48 103, U.S.A. ......... WW 403

A leaflet describing a series of on-line and offline electrostatic proofers designed for use with phototypesetting systems explains the "Matrix Electrostatic Writing Technique" and illustrates how Versatec Matrix Proofers fit into the proofing cycle of a phototypeset publication. Versatec Inc., 10100 Bubb Road, Cupertino, California 95014, U.S.A.

The HP-45 and HP-46 are pocket and desk top calculators respectively which are described in a leaflet sent to us by Hewlett-Packard Ltd. 224 Bath Road, Slough, Bucks. SLI 4DS.

Electrical and pneumatic input strip indicators are the subject of a brochure we have received. These instruments are designed in two ranges for use in control room and on-plant environments. Andrew Salanson, Penny \& Giles Lid. Mudeford, Christchurch, Hampshire BH23 4AT ..... WW 405

A six-page leaflet illustrating a comprehensive selection from their current range of electrical indicating instruments provides a selection guide for all standard ranges of Crompton instruments for applications ranging from educational and medical to industrial and military use. Crompton Parkinson Ltd, 50/52 Marefair, Northampton NN1 INY

Bahco Tools Ltd, 266B St. Ann's Road, Tottenham, London N.15, has published a leaflet containing illustrations, dimensions and cutting capacities of the range of sixteen types of pliers and nippers developed by Bahco of Sweden . .......... WW 407

We have received four brochures on new products which B Hepworth are to handle for the Hickok company of America. B. Hepworth \& Co., P.O. Box 10, Chemical Works, Kidderminster, Worcestershire.

Basic electronic systems technology .... WW 408
Fluid power teaching systems ......... WW 409
Numerical control.................WW 410
Electrenics equipment for vocational/technical programs ..............................WW 411

\section*{GENERAL INFORMATION}

The British Overseas Trade Board has published a 1973 edition of its Export Handbook for British firms trading abroad. The book is an up to date description of all the Government services available for exporters, together with mention of private agencies, addresses and a bibliography. British Overseas Trade Board, 1 Victoria Street, London S.W. 1

The Middlesex Polytechnic Prospectus for 1973-74 describes the available degree, postgraduate, diploma, certificate and short courses. Entry requirements are also listed. Middlesex Polytechnic, P.O. Box 40 , Enfield, Middlesex EN3 4SF
"Training for Company Secretaryship" is one of a series of booklets published by the Department of Employment dealing with commercial and administrative occupations prepared in accordance with the procedure laid down in the Central Training Council's Memorandum No. 7 for "training standards for occupations common to a number of industries" Her Majesty's Stationery Office, 49 High Holborn, London WCIV 6HB

Price 25 p.

\section*{Announcements}

The product range of Integrated Photomatrix Ltd, Dorchester, Dorset, is now handled in the United States by a subsidiary company, Integrated Photomatrix Inc. The new company is based at 1101 Bristol Road, Mountainside, New Jersey 07092.
"Stereo and Public Address Systems" and "Video Recording" are two courses to be presented (one lecture per week commencing in October) at Norwood Technical College, Main Building, Knight's Hill, London SE27 OTX. Applications to the Senior Administrative Officer.

Rendar Instruments Ltd have appointed Edmundson Electronic Components as their franchised distributors covering the South-East of England and the Midlands. This appointment completes a country-wide distribution network for Rendar. Customers in the South-East should place their *orders for Rendar products through Edmundson Electronic Components, 30/50 Ossory Road, London SE1 SAN. Customers in the Midlands should use the Birmingham office at \(40 / 45\) Lower Tower Street, Birmingham 19. Products include control knobs, jack plugs and sockets, switches and DIN plugs.

A five day course on "Image Processing" will be held from 12 th -16 th November at the Campus Inn, 1920 Northwestern Avenue, West Lafayette, Indiana. The course is intended to familiarize engineers and scientists with the state of the art of optical and digital image processing. For additional information contact Paul A. Wintz, Course Chairman, 605 Lingle Avenue, Lafayette, Indiana 47901 , U.S.A.

\section*{Total Communications}

\section*{Switching-centre applications: concluding part of an article on two-way information systems}

\author{
by E. J. Gargini, * M.I.E.E., M.I.E.R.E.
}

The first part of this article (September issue) concluded with a discussion of central switching systems for use in twoway information services. In this second part I shall briefly review the progress of the Rediffusion Dial-a-Program (DAP) central switching system, using it as an indication of possible future developments. \({ }^{10}\)

The basic DAP exchange frame comprises a 12096 cross-point system, using magnet-operated reed switches, for connecting any one of 36 input lines to any one (or more than one) of 336 output lines. After the introduction of a smallscale DAP installation at Thames Television, Teddington, a prototype exchange capable of operating on 36 channels, but with active equipment for 12 off-air channels and a few locally originated or two-way transmissions - was installed at Dennisport in Cape Cod, U.S.A., together with a network to reach 250 dwellings in an area of about one-third
* Rediffusion Engineering Ltd.

of a square mile. Extra lines were made available to feed extra points in half these homes. Some 160 homes are at present using this service, which became operational in August 1970. This prototype installation was undertaken as an engineering field test of the system. The results have been entirely satisfactory, both in quality of television transmission and in reliability. There has been no reported reed switch fault or failure of the distribution network, despite quite wide variation of temperature.

The Cape Cod installation includes two experimental additions which have to do with the communication concept: (1) Television cameras have been operated at two distribution points on the network, i.e. with two-way television signals passing in opposite directions on the same cable circuit; one provides local shoppers with current prices in a small supermarket, the second was used for demonstration purposes. (2) The control pair of the DAP system has been used for telephone purposes on a dedicated basis in conjunction with a two-wire version of DAP, i.e. a DAP system in which dialling and resetting signals are carried on the pair which carries the selected television signals.

Fig. 7 indicates the operation of the two-way television system in which the subscriber originated signal is sent to the exchange on a carrier frequency of twice that of output signals from the exchange. Fig. 8 shows the arrangement of the subscriber's dial unit for the two-wire DAP system.

Fig. 9 shows an application of DAP central switching now installed and working at the Case Western Reserve University at Cleveland, Ohio, U.S.A. In this system the control pair is used on a shared basis - for DAP signalling and for two-way telephonic communication between students and the video tape machine operator. The dial units can be adapted later to include a touch tone signalling system which will permit the student to control video tape machines directly.

The two-way television filtering and frequency changing equipment permits vision signal origination from a number of lecture halls which also serve as viewing centres. This equipment is of course portable and the two-way capability of the DAP system permits the use of cameras from any outlet point; thus separate feed lines for remote programme origination are not required.

Fig. 10 is an outline of a DAP installation at the Nova Park Hotel, Zurich, Switzerland, which has just become operational. This installation is a joint undertaking by Rediffusion International, Rediffusion AG, which is an independent company, and Philips. The system, when completed, will deal with a number of offair channels and 54 locally originated channels. Eighteen of the locally originated channels and initially six channels from off-air sources will be applied to the DAP exchanges for direct selection by dialling. The remaining 36 locally originated channels will be routed to indi-


Fig. 10 Dial-a-Program conference hotel system in Zurich.
vidual subscribers or groups of up to ten subscribers manually on the programme patching panel indicated. All subscribers have provision on the channel 10 position to receive a dedicated input from the patching panel, and to obtain this service the subscriber will dial a zero and use his telephone. The Nova Park Hotel is to be used as a conference centre and
conference television will be available only to delegates. A "denial" panel equipped with switches for each of 560 rooms will enable hotel staff to permit conference viewing only in rooms assigned to delegates. The switches are three-position types and control the viewing of two or three classes of programme. A programme "denial" panel is provided for placing any


Fig. 12 Total central-switching system.

one or more of the 36 DAP busbar channels into the three classes, i.e. two "denial" and one "non-denial".
A feature of the system is that any one or more channels can be pay-television channels, and a pricing panel is included for generating pay-TV pulses on selected channels at a rate determined by the value of the programme. Pay-TV meters, one per room, record the charge to be made to each guest.
Fig. 11 outlines a proposal for an alphanumeric system for airports and demonstrates the versatility of a central switching system. New alphanumeric information, i.e. flight arrival schedules, can be entered by keyboard into the central processor unit either directly or from any of the DAP outlet points using the two-way capability of the system to handle digital data.


\section*{Conclusions}

I believe the future of telecommunication lies in a total communication system of the central switching type. Fig. 12 summarizes this concept as a marriage between the telephone system of a central exchange embracing a large area and a greater number of integrated television programme and telephone concentrators or local communication exchanges.

Some \(85-90 \%\) of the wired network route in a town would be a network dedicated to individual subscribers; the remaining \(10-15 \%\) would be network dedicated to bringing information into and out of communication exchanges. The subscriber network would deal with any amount of visual data and would not need replacement with development of the visual art.

Fig. \(13^{\circ}\) indicates the simplicity of equipment in the home for an integrated system which is capable of dealing with all the features discussed in this article. Fig. 14 outlines the main components of the integrated switching centre or communication exchange.

At an I. E. E. meeting early in 1972 it was suggested that telephone microprocessors could be used to organize and link up small exchanges or concentrators to large exchanges and that this possibility becomes more attractive as the cost of integrated circuits drops.

Perhaps an organization such as the Independent Communication Authority suggested by Professor H. M. Barlow \({ }^{11}\) will be formed to consider these matters and to develop a total communication concept for the future.
Acknowledgements. I wish to acknowledge the work done by Rediffusion Engineering Ltd. and its engineers, particularly J. F. Pacey who was responsible for the installation and commissioning of the pilot Dial-a-Program projects in the U.S.A. I thank the directors of Rediffusion Engineering Ltd for permission to publish this article, although I would emphasize that the views expressed are personal and not necessarily the views of the company.

\section*{References}
10. R. P. Gabriel. "Cable TV and the Wired City". Paper presented to I.E.E., 10th Nov. 1971 (Electronics \& Power, April 1972), also R. P. Gabriel. "Experience with the Dial-a-Program System". Paper presented to I.E.E.E. North East Regional Electronics Meeting (NEREM). Boston, Mass., 2nd-5th Nov. 1971.
11. Prof. H. M. Barlow. "Telecommunications services in the U.K. Future development and overall policy", National Electronics Review, Vol.7, No.2, March/April 1971.


Fig. 14 Local communication centre.

\title{
Presenting Maintenance Information \\ Techniques developed by B.B.C. use functional diagrams and minimum of text
}

\author{
by S.W. Amos*, B.Sc., M.I.E.E.
}

The introduction of the transistor and more particularly the integrated circuit have made possible the construction of extremely compact equipments. The extent of the miniaturization possible with modern solid-state devices is well illus trated by a typical integrated circuit which contains nearly one hundred tran sistors and as many resistors - all in a package measuring lin by \(\frac{1}{4}\) in by \(\frac{1}{8}\) in! An equipment with fifty such i.cs would contain nearly 5,000 transistors: to use such a wealth of active devices would have been unthink able in the days of valves. It is practical to employ active devices in such prodigal numbers and thus to construct equipments of very great complexity because solid-state devices are inherently reliable. Nevertheless modern equipments do develop faults which must be found and corrected and this article is concerned generally with the maintenance of modern solid-state equipment and in particular with the form in which maintenance information is presented in the B.B.C.

\section*{Factors influencing the form of maintenance literature}

The following three features of modern equipment have a direct influence on the form of maintenance literature:
1. The complexity of modern equipment can be such that only the designer under stands it thoroughly, and he is unlikely to be enthusiastic about calls on his time to maintain one of his earlier designs. It follows that the equipment must be maintained by staff who do not understand its method of working in detail: they must, of course, understand or be capable of learning its operation in principle, otherwise they would be incapable of locating a fault.
2. Integrated circuits and other packaged components such as thick- and thin-film circuits cannot be repaired if they fail: they are replaced if faulty. Thus maintenance staff do not need a detailed knowledge of the internal circuitry of such devices. They must, however, know sufficient about the function of the device, its input and output voltage levels, terminating resistances etc., to be able to test it. Again, therefore, the main-
tenance man needs a general rather than a detailed knowledge of the active device.
3. Because breakdowns are rare in modern equipment, maintenance staff have little experience of tracing faults in it. When a fault does occur the maintenance man has the problem of locating the fault in an unfamiliar equipment. Thus the maintenance information must be designed to assist the rapid location of faulty areas.
Such observations prompted the B.B.C Technical Publications Section to devote some time to experiments on the form in which maintenance information for modern equipment should be presented.

It had been known for some years that maintenance men tended to rely on circuit diagrams and did not normally read associated text unless the diagram failed to give the required information. It was decided therefore to concentrate on diagrammatic forms of presentation
and to reduce text to a minimum. In early experimental forms of literature care was taken to ensure that the diagram and associated text could always be seen at the same time and the normal arrangement was for text and diagram to be on facing pages. This was an improvement on earlier layouts but still required readers to switch their attention from one page to the other in following the operation of a complex circuit. Each time the reader returned to the text or to the diagram he had to find his place and this was felt to be an undesirable interruption to the continuity of the story.

\section*{Use of functional diagrams}

There is no need to give details of the circuitry of packaged components such as i.cs but the function of such components must be indicated, otherwise it is impossible to follow the diagram.


\footnotetext{
*B.B.C. Engineering Training Centre
}

Fig. 1 Block text diagram facing circuit diagram, bath divided into functional areas.

If the maintenance man is to be able to locate faults rapidly the diagram must show clearly the interrelationships between the stages which enable the equipment to achieve its purpose. Thus the diagram must show not only the function of i.cs but also those of other stages using for example discrete components. The functions of many basic circuits are obvious to experienced maintenance men because the circuits are (or should be) drawn with a standard layout which helps rapid recognition. Typical of such well-known circuits (which can be regarded as electronic building bricks) are common-emitter amplifiers, emitter followers, long-tailed pairs; a number of others are given in BS 3939. Although such circuits may be familiar they must be recognized before their function can be appreciated and this takes a finite time: recognition of an unfamiliar layout takes an even longer time. Thus it was decided that all stages should be labelled with their function.

Great care is taken in arranging the functional blocks on the diagram to obtain a clear signal flow and whenever possible this is from left to right and from top to bottom of the diagram: main signal paths can be printed in heavy lines to distinguish them from subsidiary signal paths.

To define the boundaries of the functional stages these are printed on blue backgrounds (shown as white boxes in Figs. 1 and 2) so that the blue areas with the associated signal paths form a block diagram in which each block represents a mathematical or logical operation upon a signal. A light blue was chosen for the background colour

The illustrations in this article are taken from original drawings intended for reproduction on A3 size paper (approximately \(16 \frac{1}{2} \mathrm{in}\) by \(11 \frac{17}{8} \mathrm{in}\) ) in BBC Technical Instructions. To obtain illustrations of a size suitable for publication in Wireless World only part of each diagram is reproduced - sufficient to show the type of presentation described in the article. The backgrounds of the functional areas in Figs. 1 and 2 are printed in blue in BBC Technical Instructions but are shown as white boxes in the article. The circuit diagram of Fig. 4 is intended for reproduction in black and the explanatory notes in red but in this article the circuit is shown in white and the notes in black.
because it does not impede reading of the circuit if this is printed on it in black. An important point about this kind of diagram is that each block represents a circuit function and not an item of hardware. It could happen, for a particular equipment, that functional and hardware boundaries coincide but in general they do not. Functional diagrams aid fault location because they illustrate the division of the equipment into functions and thus give directly the information required to test any individual stage: to permit this the diagram must include terminal numbers, pin and socket connections etc., so that the input and output connections of each stage can be found on the equipment itself.

It can be assumed that a number

of basic circuits such as common-emitter amplifiers and emitter followers are so familiar to the maintenance staff that no text is necessary to explain their behaviour. Other circuits require text and this was located, in earlier maintenance instructions, on the page facing the functional diagram. To minimize the difficulty of locating the text for a particular functional block the text was also printed on blue backgrounds of the same size as those of the functional diagram and arranged in the same layout. An example of such a pair of facing pages is given in Fig. 1. It is certainly easy to find a wanted text of a functional circuit stage but this form of presenta tion is still open to the objection that the maintenance man must consult two pages and must switch his attention from one to the other in following explanations of circuit behaviour. This form of presentation can also be criticized on the grounds of duplication: the breakdown of the equipment into functional areas is shown twice, one on each page. Both difficulties can be overcome by dispensing with the block text diagram and including the text within the blue areas of each functional circuit. An example of this form of presentation is given in Fig. 2: this gives the maintenance man on one side of a piece of paper most of the information he is likely to require on the particular part of the circuit featured.

\section*{Levels of treatment}

For ease in handling, diagrams are limited in size to A3 and these are folded to A4 format for inclusion in standard folders which can be accommodated in normal-sized filing cabinets. The information which can be contained on an A3 page is limited, particularly when it is combined with text, waveforms, tables and other items of information. Thus a number of diagrams, possibly as many as 20 , are required to describe a complete equipment such as one capable of generating all the standard waveforms required to line up a picture monitor.
To break-down the circuitry into 20 diagrams without destroying the continuity of the treatment requires some thought, and the technique adopted is to present the information at a number of levels. The first diagram in the service manual (level 1) is a diagram of the complete equipment divided into its major functions which are limited to about 20 which is the maximum which can be accommodated on an A3 page with ancillary text while maintaining adequate clarity of presentation. To limit the number of functions may require some of the functions in the level-1 diagram to be complex, and at this stage it may be sufficient to label a function for example as a waveform processor (without indicating how many stages it contains). Clearly a diagram as general as this cannot contain details such as i.cs, transistors, resistors, etc.

In subsequent (level-2) diagrams these complex functions are split into simpler


Fig. 2 Text combined with circuit diagram. Fig. 3 A waveform-text diagram.
functions, these being again chosen to keep the total number of blue blocks per page to below 20 . It may be that the functions in the level- 2 diagram are so simple that the circuitry can be included within the blocks without overcrowding the diagram: if not then the functions can be subdivided further to a third level at which circuit details can be included.

In all diagrams it is essential, of course, that the functional blocks should be laid out so as to emphasize the paths of signal flow.

\section*{Waveform diagrams}

For certain types of equipment the circuit behaviour is best explained with the aid of waveform diagrams. If the account is given in conventional text with reference to separate waveform diagrams the explanations can become tedious. The usual method is to allocate letters to the edges and other significant features of the waveforms and to use these letters in referring to these features in the text. Such a technique has the disadvan-
tages already mentioned that the reader has to switch his attention between text and drawing. The repeated need of the reader to find his place in text and drawing is frustrating and wastes time: it can be avoided by using the technique employed in the functional circuit diagram, i.e. by condensing the text to a minimum, breaking it into sections and by inserting these sections at appropriate points in the waveform diagram (Fig. 3). There is then no need to label the waveform features because they can be identified by arrows. To distinguish it from the waveform the text is printed in a different colour.

\section*{Relay circuits}

Few examples of technical writing can be so boring and repetitious as detailed descriptions of the operation of a circuit including a large number of relays. It is better to provide the information in the form of a table designed to illustrate the sequence of operations. If it is possible perhaps the best method of explaining the operation of the circuit is on the circuit

diagram itself. This might be possible, for example, by stringing the individual circuits between horizontal supply lines and arranging them in the order in which they operate. By adopting this method what litde explanatory text is still needed can often be accommodated on the circuit diagram itself near the circuit in question, thus avoiding any need for separate textual description. A sample of such diagram is given in Fig. 4.

\section*{Algorithms}

One way in which it is possible to help inexperienced staff to maintain equipment is by the provision of algorithms: these are charts which esable faulty
areas to be found quickly. The charts state, for example, what signals should be present at certain points in the equipment and, if they are missing or distorted, indicate what the next test should be and where it should be applied to obtain further information on the location of the fault. Thus the algorithms give information on the logical steps in fault-finding which a skilled maintenance man would take instinctively. A sample of an algorithmic chart is given in Fig. 5.

Algorithms can be useful but the staff using them still need some experience in tracing faults. For example, an algo rithm may suggest that if a certain waveform is missing at a particular test
point then the fault lies in a specified area of the equipment. But the waveform could be absent because of a poor soldered connection at the test point itself. It is impossible to include all such possibilities in an algorithm and they are therefore of limited application.

\section*{Physical location of components}

The maintenance aids described above should enable the maintenance man to locate a faulty area in an equipment. However, additional information is needed to enable him to find any particular circuit point physically in the equipment and this is necessary of course during fault location. Functional diagrams can


Fig. 4 - relay diagram laid ou to simplify explaresion of circuit operation.


Fig. 5 An algorithmic fault-tracing chart.
provide some locational information if the terminals of transistors and the pin numbers of i.cs and of plugs and sockets are numbered. However, further information is required to enable, for example, the junction of a particular resistor and capacitor to be found physically. To this end diagrams showing the layouts of components on the printed cards are also provided and particular care is taken to identify test points on the cards.

\section*{Wiring diagrams}

For equipment which consists of a number of inter-connected units it is essential, of course, to give complete information on the inter-unit wiring. This can be in the form of a diagram or a list of connections. Probably the diagram is better, particularly if it shows the units in their correct relative positions: this simplifies transfer of attention from the printed page to the equipment itself.

\section*{Parts lists}

If a component is faulty it may still be recognizable and the type number may still be legible: all the information is then available to enable a replacement component to be obtained. Often, however, the faulty component has been destroyed (e.g. a resistor has burned out) or any markings on it have become illegible. The component cannot now be replaced until sufficient information on it has been obtained. The circuit diagram can supply some details, e.g. the resistance of resistors and the capacitance of capacitors, but this is often insufficient to enable a suitable replacement component to be obtained. Complete information on all components should therefore be included in the maintenance information. Equivalent components are sometimes satisfactory as replacements but there are some components for which replacements must be precisely the same type as those used originally.

\section*{Conclusion}

The methods outlined in this article have been introduced into B.B.C. Technical Instructions over the last three years and are regarded by the maintenance staff as a considerable improvement over earlier methods of presenting maintenance
information. In particular the reduction in the volume of conventional text and the introduction of the functional diagrams have been welcomed. Experiments in presentation will continue but it is anticipated that changes will be confined to details in the immediate future.

\section*{Books Received}

Electronic Maintenance Management contains the contributions made to the 1973 Symposium of the Society of Electronic and Radio Technicians held at the University of Nottingham earlier this year. Subjects covered range through maintenance philosophies, technical documentation and design requirements to personnel organizations and careers. Speakers at the symposium represented all sectors of the industry from large to smal specialized companies. Titles of the 21 papers contained in the proceedings include Education and Training for Maintenance Management, The Economics of Servicing, Training in Fault Diagnostic Techniques, The Effect of Service on Design, The Use of Algorithmic Fault Finding Guides, The Maintenance Task on Commercial Computers - A Different Approach, and The Need for a Standard Format of Maintenance Data for Electronic Equipment. Price £5 (incl. p \& p.). Pp. 189 plus unpaginated papers (3). Society of Electronic and Radio Technicians, Faraday House, 8-10 Charing Cross Road, London WC2H 0HP.

Search the Solar System by James Strong discusses the future role of unmanned interplanetary probes. Emphasis has now been placed on the continued exploration of the Solar System by probes similar to the "Mariner" reconnaissance of Mars and Venus and the "Pioneer" probe now on its way to Jupiter. Because every planet presents a fresh set of problems, various types of space probe will be necessary. Some will be purely reconnaissance orbiters while others will soft-land sophisticated
capsules that will search for evidence of life on the surface by remote control. The author discusses ways of exploring hot planets, like Venus and Mercury, and how fast- or slowmoving comets can be intercepted. He also describes how to control a television-guided mobile probe, special balloon probes and radar satellites, how to explore the rings of Saturn, and describes a new way of maintaining continuous radio communication between Earth and a planetary surface anywhere in the Solar System. The book also describes the latest techniques for sending fast probes to the Outer Planets and a "kamikaze" probe to take close-up pictures of the Sun. Price \(£ 3.25\). Pp. 160. David \& Charles (Holdings) Ltd, South Devon House, Newton Abbot, Devon

Included in recent additions to the list of books in the Foulsham-Tab series and published by W. Foulsham \& Co. Ltd, Yeovil Road, Slough. SLI 4JH are:

Radio Control Manual - Systems Circuits and Construction by Edward L. Safford. Price £1.25. Pp. 190.
Audio Systems Handbook by Norman H. Crowhurst. Price £1.25. Pp. 189.

New IC FET Principles \& Projects by K.W. Sessions and D. Tuite. Price £1.10. Pp. 160.

Simple Transistor Projects for Hobbyists and Students by Larry Steckler. Price \(£ 1.25\). Pp. 192.

Video Tape Production \& Communication Techniques by Joel L. Efrein. Price £1.30. Pp. 252.

\section*{Circuit Ideas}

\section*{Avoiding power supply hum}

Units such as a radio tuner or tape recorder feeding into an amplifier normally require their own separate power supplies, though it is often more convenient if such feeders, with their relatively small requirements, could take their supplies from the amplifier itself. This cannot be achieved without the earth line between feeder and amplifier being shared by both the power supply and the signal. The result of this arrangement is considerable hum due to positive feedback.

This can be relatively easily overcome. The power supply used by the amplifier will almost certainly be of higher voltage than that needed by the feeder, and thus a "potentiometer" can be used to reduce the voltage. If this consists of a constant-current source in series with a constant-voltage sink, and is close up against the feeder, then feedback is eliminated as there will be no current fluctuations in the power supply line.

The first circuit (left) provides power for a small cassette recorder. It supplies up to 150 mA at 7.5 volts from supplies varying from 12 to 24 volts, and gives completely hum-free service.


In the more difficult case of a high-quality tuner (e.g. Nelson-Jones), together with stereo decoder, the second circuit has been used with similar success, even at the end of eight feet of cable.
With its exceedingly low output impedance, this circuit not only eliminates all likely sources of feedback, but also provides the high degree of smoothing required by this tuner.

\section*{Symmetrical power supply}

Shown is a simple power supply having symmetrical outputs and overload protection such that if a heavy load or short circuit is applied to either output both switch off rapialy. Each transistor derives its forward bias from the opposite supply rail and while the transistors are in saturation an increasing load will cause the output to fall under the regulation of the transformer. With a further increase in load the transistors come out of saturation and eventually remove each other's forward bias. If the trip circuit
is used with an active power supply having negligible output resistance then switch off occurs solely to the transistors coming out of saturation. The zener voltage must be between \(V_{o}\) and \(2 V_{o}\) and the value of \(R\)
\[
\leqslant \frac{h_{F E}\left(2 V_{o}-V_{Z}-0.7\right)}{I_{L}}
\]
where \(I_{L}\) is the maximum load current required.
L. D. Thomas,

Post Office Research Department.



With the values shown the constantcurrent supplied is about 100 mA of which 10 to 15 mA are "sunk" by the 741 . The circuit requires careful setting up as the current that can be sunk is severely limited. Coarse adjustment of the current source is achieved by altering \(R_{1}\), the BD136 emitter resistor, fine adjustment by \(R_{2}\) Giles Hibbert,
Blackfriars,
Oxford.

\section*{Resistance-to-voltage converter has low output impedance}

Driving a constant current through an unknown resistance \(R_{x}\) yields a voltage across it proportional to the unknown's value. This commonly used method provides a linear conversion function, but the

signal source has an output resistance equal to \(R_{x}\). Better drive capability is available by using an operational amplifier in the inverting configuration, resulting in closed loop gains of \(-R_{x} / R_{i}\) when the unknown value functions as the feedback resistor. Constant input \(V_{\text {ref }}\) then results in output voltage directly proportional to the resistance value and a power of ten scaling factor selected by proper choice of \(R_{i}\) and \(V_{\text {ref }}\). Output voltage from the amplifier has a very low source resistance, which approaches zero as \(R_{i}\) is set at higher values and loop gain decreases.
David R. Schaller,
Milwaukee,
Wisconsin.

\section*{World of Amateur Radio}

\section*{Should there be a U.K. novice licence?}

The U.K. is one of the very few countries with a large amateur population (total of Class A and B licences has just passed the 18,000 mark) that eschews "incentive licensing" and any form of novice or beginner's licence. Once anyone has qualified for a Class A or B licence no restrictions are placed on their activities or whether this includes any element of "self-training". As an individualist such freedom from official pressures seems wholly admirable - but as one concerned with the future development of the hobby this absence of incentives seems puzzling. A British "beginner's licence" was announced in March 1968 but did not appear to be part of any fully thought-out scheme and was never implemented.

One of the latest countries to introduce such a system is Norway where a new novice-type Class \(\mathbf{B}\) licence is now being issued. This is valid for two years only and is not renewable, is for c.w. operation with a maximum input of 15 watts and has the prefix LB to distinguish these stations from the Norwegian Class A stations which use the LA prefix. Such facilities encourage newcomers to become proficient in Morse operation by learning while using. Should the U.K. look again at this now well tried system of encouraging newcomers to become experienced in h.f. Morse operation? Or is it accepted that the day of the Morse key is now almost done?

\section*{Communications receivers now 40 years old}

Although it could be argued that "communications receivers" are as old as radio, the type of receiver which this term usually defines really emerged in the early thirties with the coming of lownoise single-signal-superhets. A major step forward came in 1931 when James Lamb of A.R.R.L. showed how the crystal i.f. filter - developed in the U.K. by Dr J. Robinson in 1929 for his broadcast "stenode" receiver - could be put to extremely good use for amateur operation. One of the first receivers using such a filter and intended for amateurs was the National FBXA of 1934 - and
this was a later version of the FB7 receiver of 1933, regarded by some amateur historians as perhaps the first true "communications receiver". During the next five years a whole string of such receivers appeared and pushed aside the "straight" regenerative receivers: the HRO Senior in 1936; the Hammarlund Comet Pro and Super Pro; the RME69; the Tobe Deutschmann kit; the wide range of Hallicrafters from the \(£ 7\) Sky Buddy to more advanced sets with crystal filters. At about the time when such receivers began appearing on the British market in 1936-37, the only home product was possibly the \(£ 20\) Evrizone single-signalsuperhet. A sign of the change was the use by the leading British station of the 1936 BERU contest of a Comet Pro - but the Australian winner still used a straight \(0-\mathrm{v}-2\) receiver. Because of their widespread wartime use, the HRO and the slightly later RCA AR88 (first marketed in 1940-41) remain as outstanding examples of the early days of communications receivers: in how many other branches of electronics can equipment built over 30 years ago still prove capable of performing well even in comparison with modern equipment?

\section*{Coming Soon}

The second Midland National Amateur Radio \& Electronics Exhibition at Granby Halls, Leicester on October 25-27 will feature a fully equipped amateur station, GB3ARE, plus a reproduction of the 1913 club station of the Derby and District society. A film theatre will show films of interest to amateurs and a Tombola stand is being run on behalf of the Radio Amateur's Invalid \& Bedfast Club. Trade stands look like being fully booked.

One of the very popular Racal Amateur Radio Club Junk Sales - at which an unusually wide range of ex-professional equipment usually changes hands - is being held at St Sebastian's Hall, Nine Mile Ride, Crowthorne, Berkshire on Saturday, October 27 at 2 p.m.

The South East Counties H.F. Convention on Sunday, November 18 from 11 a.m. until 7 p.m. at the Airport Hotel, Crawley, Sussex, will include trade stands, club stands, informal lectures and an operational station.

A dinner for direction-finding enthusiasts which it is hoped will be the first of an annual event is to be held at "The Chicken in the Basket", Benson, between Oxford and Wallingford on Friday, November 16.

\section*{More courses for would-be amateurs}

Additions to last month's list of places where evening courses are being run for would-be amateurs - based on information supplied by the R.S.G.B. include:
London and the Home Counties: Acton, Bedford, Borehamwood, Brentwood, Chingford, Croydon, Harlow, Harrow, Highgate, Ilford, Islington, Princes Risborough.
Provinces: Aldridge, Staffs, Bangor, Co. Down, Birkenhead, Bridgend, Glam. Brighton, Bury, Chesterfield, Glasgow, Grantham, Grimsby, Loughborough, Newport, Mon., Oldham, Perth, Plymouth, Portsmouth, Stoke-on-Trent, Wolverton and Wombourne (near Wolverhampton).

\section*{In Brief}

A recent BBC Radio 4 programme ("In Touch") for visually handicapped listeners included an explanation and taped demonstrations of amateur radio operation by Philip Storey, G3ZGG. He said there are now about 50 blind amateurs in the U.K. and advised those interested to gain experience by listening on a communications receiver; where necessary assistance may be obtained from the Radio Amateur Invalid \& Bedfast Club (Hon. Secretary Mrs Frances Woolley, G3LWY, Woodclose, Penselwood, Wincanton, Somerset) . . . What is thought to be the longest-distance contact ever made on "Top Band" ( 1.8 MHz ) was made last July between Tokuro Matsumoto, JA7AO and VP8KF on the Falkland Islands. The Japanese amateur has also contacted Fred Laun, LU5HFI (formerly HS5ABD) in Cordoba. Argentina . . . The Harlow and District Mobile Rally is to be held at Netteswell Comprehensive School, Harlow, onSunday, September 23 (talk-in stations on 144, 3.5 and 1.8 MHz ) . . . The prefix DT is being used by some DM stations until the end of the year to mark the 20th anniversary of amateur licensing in the German Democratic Republic . . . Walter Turner, GW3YPH, of Pontypridd was electrocuted while putting up an aerial in his back garden not far from an overhead electric cable . . . By the end of August, Oscar 6 has completed 4000 orbits. British and French amateurs have made contact through Oscar 6 with KL7MF at Anchorage, Alaska, at distances of over 7000 km . F9FT in Rheims has made more than 2500 contacts through the amateur satellite.

PAT HAWKER, G3VA


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}

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* Constant acts as last entry in a calculation.
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* Calculates to 8 significant digits, with exponent range from \(10^{-20}\) to \(10^{79}\).
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\section*{New Products}

\section*{Two-tone generator}

Racal Instruments have introduced a two-tone generator, Model 9063, which complements the r.f. synthesized signal generator system introduced earlier this year. Designed primarily as a modulation source, the 9063 provides an aid to s.s.b. servicing and alignment.

An entirely sclf-contained instrument covering the 10 Hz to 100 kHz frequency range, the 9063 provides a stable synthesized tone referenced to an internal crystal standard and a second tone derived from a tuneable oscillator. The two tones may be used independently or combined, the intermodulation distortion being less than -70 dB . Fully remotely programmable using t.t.l. logic, the instrument provides outputs from \(100 \mu \mathrm{~V}\) to 10 V with low hum and spurious content. The 9063 is 88 mm ( 3.5 in ) high, 483 mm ( 19 in ) wide and 406 mm ( 16 in ) deep and weighs 13.7 kg (30lbs). Racal Instruments Ltd, Duke Street. Windsor, Berks SL4 1SB.
WW 316 for further details

\section*{50 MHz portable oscilloscope}

The dual-trace oscilloscope, model D75 from Telequipment is a light-weight portable oscilloscope with a vertical sensitivity of \(5 \mathrm{mV} /\) div. on both channels over the full 50 MHz bandwidth. The sensitivity can be increased to \(1 \mathrm{mV} /\) div. at all frequencies up to 15 MHz by the operation of the \(\times 5\) gain switch. Both vertical channels can be used independently or in alternate, chopped, added or differentia! modes, and the vertical signal delay which
is incorporated allows the leading edges of signals to be observed.

The horizontal deflection system consists of a dual timebase which provides normal sweep, mixed sweep, sweep intensifying, sweep delaying and single shot facilities. The fastest normal sweep speed of \(100 \mathrm{~ns} /\) div. can be increased to \(10 \mathrm{~ns} /\) div. by switching on the \(\times 10\) sweep magnifier. Time measuring accuracy is \(\pm 3 \%\) with the magnifier off, \(\pm 6 \%\) with the magnifier on. A useful extra feature is the "trace locate" button which reduces vertical and horizontal deflection so as to present an on-screen display and free runs the main timebase.

A built-in calibrator provides square wave test outputs on the front panel of \(30 \mathrm{mV}, 300 \mathrm{mV}\) and 3 mA peak-to-peak with an accuracy of \(\pm 1 \%\) at a nominal frequency of 1 kHz . The c.r.t. is a singlegun mesh tube, operating at a potential of 15 kV . The U.K. list price of the D75 is \(£ 420\) excluding v.a.t. The weight is 25.5 lb and the dimensions are \(5 \frac{3}{8} \mathrm{in}\) high, 15 in wide, and \(18 \frac{3}{4}\) in deep. Tektronix U.K. Ltd, Beaverton House, P.O. Box 69, Harpenden, Herts.

WW 307 for further details

\section*{Metal detector}

The Contil-Voll metal detective, marketed by West Hyde Developments, is pocket size, hand held, very light (under \(\frac{1}{2} \mathrm{lb}\) ) and so easy to operate that a child can use it. It is held in either hand and a small knurled knob is turned by the thumb until a red indicator light

goes out. It is then ready for use: any nearby metal turns the light on again. The Contil-Voll metal detective reacts to all metals and finds them through any other substance. To quote West Hyde, "just think of the possibilities; this means the metal clip on your pen through your jacket; the money in your pocket; a gun in your belt! It will find a paper staple at \(\frac{3}{4}\) in and a paper clip buried under a pile of papers 2 in thick. But think of the practical possibilities in a factory or at home!" Made in tough impact-proof plastic and using a single 9 V battery as the power source, it is priced at \(£ 22.85\) plus v.a.t. West Hyde Developments Ltd., Ryefield Crescent, Northwood Hills, Northwood, Middx HA6 1NN.
WW 308 for further details

\section*{Electronic watches}

Solidev Ltd announce a range of solid quartz controlled electronic watches. They have no hands and no dials. Instead, a simple black face registers time and date by l.e.d. display. Control is by push button.

There are several models available in both men's and ladies' styles. All feature gold bracelets. All are accurate to within 1 digit over the course of a year, and cost \(£ 180\) r.r.p. Solidev Ltd, Edison Road, Elms Industrial Estate. Bedford MK410HG.
WW 319 for further details

\section*{Trip amplifiers}

Rotraco Systems Ltd have developed a range of trip amplifiers which are compatible with their Model 4 and 4R monitoring and alarm systems. The trip amplifiers are intended for operation from signal sources providing \(0-2.5 \mathrm{~V}, 4-20 \mathrm{~mA}\), or direct from thermocouples, resistance thermometers, thermistors, strain gauge pressure transducers, tachometers etc. They can be combined with alarms to provide alarm and trip systems for plant and process control applications. Facilities are also provided for an indication of temperature, pressure etc. either on a common indicator or individual indicators. The indicator unit can also be used to inject a signal to check the trip setting.

The basic trip amplifiers carry two trip channels with controls for set point and differential but other versions are available e.g. one channel with two trip settings, switch selection of trip high or low, fixed differential etc. A light emitting diode indicates whether or not a channel has tripped. The setting and differential potentiometers of temperature trip amplifiers are calibrated directly in degrees centigrade and for other parameters they

are calibrated in percentage. Rotraco Systems Ltd, Garden Street, Darlington, Co. Durham DL1 1QR.
WW 314 for further details

\section*{Liquid level control units}

The ELC2 electronic level control produced by Gearing \& Watson (Electronics), is used to control the level of liquids where the conductivity varies with level, such as water, milk, sewage etc. Two level probes may be used, so that the level is controlled between the two probes; for example a container may be filled to the upper level probe but the level may fall to the lower probe before the control operates to restore the level to the upper probe again. The ELC2 is designed for "fill" applications, the ELC24 is available with a reverse action.

The ELC2 uses modern semiconductor techrology, the level comparator being an integrated circuit, while the load switch is a triac; thus no contact maintenance is required. The probe circuit is insulated by a double wound mains transformer from the supply for complete safety. The unit is encapsulated for protection and can operate in high ambient temperatures. Electrical connections are by "Faston" connectors. Dimensions \(63.5 \times\) \(51 \times 108 \mathrm{~mm}\). Gearing \& Watson (Electronics) Ltd, Birch Close, Eastbourne, Sussex BN23 6PE.

\section*{WW 306 for further details}

\section*{Logic analyzer}

Logic circuit ánalysis by the "freezing" and display of the states of any 32 bits in a stream is the function of a new instrument by Hewlett-Packard, the Model 5000 A . A light-emitting diode array indicates the "up" or "down" states of the 32 selected bits, which are selectable by thumbwheel switches from a stream of information in relation to a trigger signal (before it as well as after). A fault con-
dition can be made to trigger the display. The instrument will cope with clock frequencies up to 10 MHz and is compatible with all logic families. L.e.ds display the state of the input signal, and if the probe is not making contact, the fact is again indicated by l.e.d. Hewlett-Packard Ltd, 224 Bath Road, Slough, Bucks SL1 4DS. WW305 for further details


\section*{Glass miniature trimmers}

Voltronics are now manufacturing a range of glass dielectric trimmers up to 40 pF capacitance, but 40 per cent shorter in length than the standard MIL-C 14409 C capacitor. Two mounting styles, for vertical or horizontal mounting, are available. Both are sealed, providing protection against pressures up to 40 p.s.i. to keep out dust, moisture and encapsulents. The non-rotating piston construction permits more linear tuning, no capacitance reversals, high \(Q\), longer life, low constant inductance and a high selfresonant frequency. Suvicon Ltd, Hagley House, Hagley Road, Birmingham B16 8QW.
WW 304 for further details

\section*{Hall effect d.c. motors}

Communication Technology have announced a range of constant speed brushless d.c. motors for instrumentation and recorder applications. Incorporating Hall effect elements to control the coil current by sensing the rotor position, the PHM series can be programmed to operate at selected motor speeds or for reverse rotation. By the addition of an external switching circuit, multispeed operation and forward or reverse rotation can be arranged without the need to use a separate gear box for this purpose.

The rotator is external to the fixed coils and as there are no brushes or a commutator, as in a conventional d.c. motor, friction is greatly reduced and
no electrical noise is generated. The use of special copper sleeve bearings is said to ensure a minimum of motor noise and long life. Communication Technology Ltd, 279 Addiscombe Road, Croydon, CRO 7HY.
WW 303 for further details

\section*{Audio mixers}

Neltronic (UK) Ltd is to enter the audio market with a range of audio mixers, with full mixing input channels and frequency response 20 Hz to 20 kHz . In size and sophistication the range covers many requirements but units can be designed specially to fulfil customers' needs. The equipment is made in the UK, to a design which employs up-todate technology and components to bring the price of the standard unit, model \(6 / 2\), to under \(£ 900\). For the standard unit, the mixer features switching for one microphone and one line input, a sensitivity -80 to \(\pm 10 \mathrm{dBM}\) in 10 dB steps, filters covering low, mid and high frequencies and horizontal faders. Cueing to internal speaker from all channels and all outputs is available with independent reverberation, selectable mix, 2 reverberation units for stereo operation, switchable to main output, a fold-back output for mixed signals and a p.u. unit with R.I.A.A. equalization. Two v.u. meters switchable to every channel and all outputs; pan between 1 and 2 on every channel. Outputs 2 group, 2 foldback, 2 reverberation and one monitor. Neltronic (UK) Ltd, 442 Bath Road, Slough SL1 6BB, Bucks.
WW302 for further details

\section*{Noise generator cards}

Mariufactured by Elgenco, Inc., U.S.A., specialists in noise generators, the Series 3600 noise generator cards cover the range of 10 Hz to 5 MHz with an output level of 3 V r.m.s. open circuit. A dynamic range of \(3.5: 1\) peak to r.m.s. is provided. Output impedance is \(600 \Omega\) or \(200 \Omega\) depending on the upper frequency specification, with other output impedances optionally available. The amplitude probability distribution is Gaussian.

Cards are available with specified uniformities of \(\pm 0.5 \mathrm{~dB}, \pm 1 \mathrm{~dB}, \pm 2 \mathrm{~dB}\) and \(\pm 3 \mathrm{~dB}\) for many frequency ranges between lower frequencies of \(10 \mathrm{~Hz}, 20 \mathrm{~Hz}, 50 \mathrm{~Hz}\), 200 Hz and 5 kHz and upper frequencies of \(20 \mathrm{kHz}, 50 \mathrm{kHz}, 100 \mathrm{kHz}, 200 \mathrm{kHz}, 500 \mathrm{kHz}\), \(1 \mathrm{MHz}, 2 \mathrm{MHz}\) and 5 MHz . Size is \(4 \frac{1}{2} \times\) \(6 \frac{1}{2} \times \frac{71}{8} \mathrm{in}\) and weight approximately \(70 z\). Single unit prices range from \(£ 87\) to \(£ 285\) (excluding duty and v.a.t.), lower prices in quantity. Lyons Instruments Ltd, Hoddesdon, Herts.
WW313 for further details

\section*{Three-pen recorder}

The Model 303 "Dial-a-Span" three-pen recorder by Chessell Ltd is designed to provide good flexibility. The ranging facility is controlled by front panel thumbwheel selectors to give 450 spans per channel from 1 mV to 99 V plus 1999 datum shift settings permitting at least 10 times span suppression or elevation on all ranges, with a constant input impedance of \(10 \mathrm{M}:\)

The calibrated datum shift facility
allows 1 metre chart width resolution on a 100 mm chart. Other features include high speed pen servos and 10 speed electronic chart drive. The Model 303 recorder is available in two forms: a free-standing model with integral carrying handle/tilt stand and a 19 in rack mounting version. Chessell Ltd, Broadwater Trading Estate, Southdownview Road, Worthing, Sussex BN 14 8NL.
WW 311 for further details


\section*{Low-cost TV system}

Pye Business Communications has introduced a low-cost television system the Philips" "Mini-Studio". Designed for educational and industrial training purposes, the equipment, constituting a basic studio, costs \(£ 1,450\). The system includes two cameras with various lenses (zoom, standard microscopy), tripods, camera fixings, headsets, telecine,
mounting rack with three 10 cm monitors and control unit, microphone and audio mixer with all necessary cables, stands and connections. The TV equipment can be used with either a video tape recorder (from £150) or video cassette recorder with professional facilities (from £750). Pye Business Communications Ltd, Cromwell Road, Cambridge. WW 310 for further details


\section*{D.I.P. heat pipes}

Small flat pipes for cooling flat packs and d.i.p. devices are the latest designs from Jermyn Manufacturing. Each pipe is \(0 . l\) in thick and 0.25 in wide and will easily fit under d.i.p. packs. The flat surface of a d.i.p. heat pipe makes it easy to add or remove heat and due to the isothermal characteristics of heat pipes all devices will be maintained at the same temperature.

The di.i.p. heat pipes can be fabricated in matrix configurations with the heat
pipe tails terminating in a cold wall and on a \(4 \times 7\) in matrix up to 80 d.i.p. packs can be maintained within \(1^{\circ} \mathrm{C}\) of each other.

The normal temperature of operation is \(20^{\circ} \mathrm{C}\) to \(150^{\circ} \mathrm{C}\) and a 6 in long d.i.p. heat pipe will handle 18 watts at \(100^{\circ} \mathrm{C}\) and with a weight of only 10 g these components are suitable for airborne applications. Jermyn Manufacturing, Sevenoaks, Kent. WW 312 for further details

\section*{Modular bench supply}

Gardners Transformers have introduced a power supply system for industrial and educational laboratory applications where either a.c. or d.c. may be required at power levels up to several hundred watts. The complete power supply system comprises a basic a.c. module, ACO , and four d.c. modules. The a.c. modules provide a continuously variable a.c. output, fully isolated, with coarse and fine controls, together with meter monitoring. There are three output ranges: up to 70 V at 12 A ; up to 140 V at 6 A ; and up to 280 V at 3 A .
The incorporated isolation transformer ensures that the user and valuable test equipment are protected against the risks associated with testing circuits connected
to mains earth. This feature, coupled with the usual flexibility of the system as a whole, should be valuable where students or inexperienced personnel are involved.

Three of the d.c. modules are rectifiercapacitor units matching the three output ranges of the a.c. module. The fourth d.c. module, which can be used in conjunction with any of the other d.c. modules, is a filter and protection unit offering very low output ripple levels in addition to full short-circuit and overvoltage protection. The complete system is housed in two instrument cases which may be used on the bench or mounted in standard 19in racks. Gardners Transformers Ltd, Christchurch, Hants. WW 309 for further details


\section*{Time delay modules}

The time delay modules TM and TD by Keyswitch can be supplied for a.c. or d.c. applications where timed delay periods of 2.5 to 300 s are required. The units are

"non-blip", and the delay period is set by potentiometer adjustment. This potentiometer can be included on the discrete unit, or wired from a remote situation. At the end of the set delay period (which is initiated by connection of the supply) the timer will deliver an output.

The TM timer incorporates a Keyswitch MS relay with changeover contacts rated at 2 A . The TD version is designed for use with an external relay, and at the end of the timed interval the TD timer output is supplied via an integral s.c.r. circuit. The solid state switch rating is \(300 \mathrm{~V}, 10-800 \mathrm{~mA}\).

Both timers have 120 ms reset time, and are supplied in polypropylene casings 1.3 in \((33 \mathrm{~mm})\) wide \(\times 1.2\) in \((33 \mathrm{~mm})\) deep, overall height above socket 2.025 in ( 51 mm ). Gothic Electronic Components, Beacon House, Hampton Street, Birmingham 19.

\section*{WW 315 for further details}

\section*{Ceramic filters}
10.7 MHz centre frequency ceramic filters, type CFS 107M, are offered by Toko (UK) Ltd. The five ranges available have centre frequencies between 10.64 and 10.76 MHz , each with an accuracy of \(\pm 300 \mathrm{kHz}\) at 3 dB and \(\pm 600 \mathrm{kHz}\) at 20 dB . Insertion loss is not more than 6 dB , and impedance \(330 \Omega \pm 15 \%\).


The performance parameters are closely specified, and include a centre frequency shift within \(\pm 150\) p.p.m. per \({ }^{\circ} \mathrm{C}\) in the temperature range -10 to \(+60^{\circ} \mathrm{C}\).

Delivery times are short and prices competitive for these type CFS 107 M ceramic filters claim the suppliers, Toko (UK) Ltd, Shirley Lodge, 470 London Road, Slough, Bucks SL3 8QY.
WW301 for further details

Each section under the title of Solid State, is devoted to the new semiconductor products offered by one manufacturer or distributor. The type number and device title is given in bold type, followed by a brief description of features or application. The section is terminated with the address of the company together with reader reply card numbers associated with the device numbers or types.

\section*{Announced from Guest International} Ltd:
SH730. Sample and hold amplifier manufactured by Hybrid Systems Corporation. This is characterized by a linearity of \(0.01 \%\) and the capability of acquiring a \(\pm 10 \mathrm{~V}\) to this accuracy in less than \(1 \mu \mathrm{~s}\). The mode control is t.t.1./d.t.t. compatible and the droop rate of \(5 \mathrm{mV} / \mathrm{ms}\) can be improved by the addition of an external capacitor which, however, lengthens the acquisition time.

4403 and 4440 Red-Lit high brightness l.e.ds are Gallium Arsenide devices having a luminous intensity of 1.2 mcd at 20 mA and a power dissipation of 200 mW . The 4440 is a lower cost version with a luminous intensity of 0.8 mcd at 20 mA . Both can be soldered directly to a p.c.b. or mounted in a panel with a snap-in mounting clip.

DAC 328-4-BCD 4 decade b.c.d. current output digital-to-analogue converter, packaged in a \(2 \times 2 \times 0.4\) in module. It is t.t.1./d.t.l. compatible and operates from a standard \(\pm 15 \mathrm{~V}\) power supply.

Industrial Electronic Components Division, Guest International Ltd, Redlands, Coulsdon, Surrey CR3 2HT.
WW 350 sample /hold amplifier
WW 351 high brightness l.e.ds
WW352 4 decade b.c.d.

\section*{Announced from Mullard:}

TAA320A voltage level detector is an i.c. for use in flame control systems, radiation detectors, timers, thermostats and liquid level detectors. It is a development of the audio amplifier TAA320. Operating with an input current of only 1 pA , it will produce an output of 60 mA at 20 V when a predetermined input threshold voltage is exceeded.

SAJI10 frequency divider i.c. is suitable for electronic organ applications and will produce seven different notes. Thus, twelve i.cs with oscillators to generate the fundamentals will provide all the notes required in an electronic organ. Combining outputs via resistor networks produces an increased range of harmonics and facilitates the synthesis of a wider number of tones.

TCA420 i.f. amplifier for use with f.m. receivers, contains a four-stage i.f. amplifier/limiter and a symmetrical quadrature detector providing a high degree of a.m. rejection. It will also supply an output for a tuning meter, and an automatic stereo inhibit switch when the signal drops below a predetermined value. The switch hysteresis can be adjusted to prevent its continual operation by small signal changes. Inter-station muting is provided and to assist tuning the TCA420A also has "side response" damping.

Mullard Ltd, Mullard House, Torrington Place, London WC1E 7HD.
WW353 voltage level detector
WW354 frequency divider
WW355 i.f. amplifier

Announced from Burr-Brown International Ltd:
SHC23 hybrid sample /hold amplifier has a guaranteed dynamic non-linearity of less than \(\pm 0.01 \%\). The addition of an external capacitor enables optimization of acquisition time and hold time. The device is packaged in a TO-8 can and is available with an operating temperature range of \(0^{\circ} \mathrm{C}\) to \(+70^{\circ} \mathrm{C}\).

3506J and 3508J high-gain wideband operational amplifiers. The model 3506J has a small signal bandwidth of 12 MHz and a slew rate of \(7 \mathrm{~V} / \mu \mathrm{s}\) and is internally compensated for stability at all gains, including the unity gain voltage follower configuration. Model 3508J has a gainbandwidth product of 100 MHz , a slew rate of \(35 \mathrm{~V} / \mu \mathrm{s}\) and is internally compensated for all gains greater than five. Both units have an open loop gain of 103 dB into \(2 \mathrm{k} \Omega\), a common mode rejection of 100 dB , bias current of 25 nA and a differential input impedance of \(300 \mathrm{M} \Omega\).

3507J high-slew operational amplifier has a slew rate of \(120 \mathrm{~V} / \mu \mathrm{s}\), a settling time to \(0.1 \%\) of 200 ns and a gainbandwidth product of 20 MHz . For gains greater than 3 the roll-off is 6 dB /octave and by adding a single external 20 pF capacitor the 3507J can be stabilized at unity gain.

3505J fast settling operational amplifier will settle to \(0.1 \%\) in 300 ns , gives a slew rate \(30 \mathrm{~V} / \mu \mathrm{s}\) and a gain-bandwidth product of 12 MHz . The amplifier is stable at all gains down to unity. without the need for external compensation.

4310 true r.m.s.to-d.c. converter uses the thermal conversion technique. The heart of the device is a pair of monolithic chips, each containing a resistor-transistor element. The function of the converter is first input voltage to heat, followed by conversion of the heat into a current change and finally from the current change to a d.c. voltage equivalent to the true r.m.s. value of the input voltage. Two versions are available, the 4130 K characterized by midrange accuracies of \(0.05 \%\) and non-linearity of 0.4 mV , and the 4130 J providing an accuracy of \(0.1 \%\) and non-linearity of 1.0 mV . Minimum bandwidth of both units is 40 Hz to 100 kHz and a 10 MHz minimum upper frequency limit for \(2 \%\) accuracy. Crest factors from 5:1 up to 100:1 maximum and the ability to accept from 100 mV r.m.s. to 2 V r.m.s. with peaks up to \(\pm 10 \mathrm{~V}\) are also typical. Fully protected from over-voltage, the 4130 also features an input impedance of \(10 \mathrm{k} \Omega\), a settling time of 4 s or less, external adjustments for gain, offset voltage and l.f. cut-off.

UAF11/15 series and UAF21/25 series hybrid active filters. Both of these are produced in di.i. packages and can be externally tuned for gain, frequency and \(Q\) over their specified ranges by adding four resistors. The basic filter utilizes the state variable principle in which low-pass, bandpass and high-pass responses are simultaneously available as outputs from a single 2-pole filter element. Complex responses can be realized for almost any filter function by cascading units. Full power bandwidth for the UAF11/15 series low pass output is 10 kHz for \(\pm 10 \mathrm{~V}\) signal ranges and is useable at frequencies up to 100 kHz for \(\pm 1 \mathrm{~V}\) signal ranges. The UAF21/25 series have a full power bandwidth of 100 kHz at the low-pass output for \(\pm 10 \mathrm{~V}\) signal ranges and is useable up to 1 MHz for \(\pm 1 \mathrm{~V}\) signal ranges.

Burr-Brown International Ltd, 25A King Street, Watford WDI 8BT.

WW356 hybrid sample/hold amplifier WW357 wideband operational amplifiers WW358 high-slew operational amplifier WW359 fast settling operational amplifier WW360 true r.m.s.-to-d.c. converter WW361 hybrid active filters

\title{
Real and Imaginary
}
by "Vector"

\title{
Odd Ode \\ (with apologies to Cyril Fletcher et al)
}

This is the tale of Phil A Ment
Whose pa, a scientific gent, Subjected himself religiously
To r.f. in large quantity
"For", quoth he, "This dosage will
Prove as effective as the Pill."
- Alas for Family Planning, when His spouse conceived a sudden yen - Quite devoid of rhyme or reason

For strawberries right out of season
Which forced papa in chagrined terror To admit experimental error
With his "Letter to Nature" finishing in
A convenient waste-paper bin.
But the r.f. currents by some means
Had gingered the paternal genes
For baby Phil 'twas plain to see
Was born an infant prodigy
At six months old the little tot
Instead of lying in his cot
And practising his coo and drool
Was learning Fleming's Left-hand Rule
And by the tender age of three
Had mastered trigonometry,
Sine waves and alternating forces
And eddy currents and their losses
Capping this tour de force sans fuss
With differential calculus -
By five he'd found a grievous flaw In Einstein's monumental Law.
His school career through "prep" and "high"
And likewise university
Was lustrous with "distincts" and "hons"
(He frequently advised the dons.)
Thus at the age of twenty he
Acquired a king-sized Ph.D.
Young Phil, his banner thus unfurled
Emerged into the outside world
Turning his back on cloisters he Elected to join our industry
And soon the rash youth deftly nabs
A job at the Gargantuan Labs.
His team-mate here was Humphrey Naild
A redbrick Bachelor of Science (failed)
Who thought that Ohm was where the 'eart is
And only shone at office parties.
Phil, being an ambitious lad And anxious to acquire a wad Of doubloons, burnt the midnight tapers And churned out reams of Learned Papers

Which brought him references in flocks
But nothing else, for Learned Socs.
Adopt the parsimonious stance
That honour is adequate advance -
A commodity which honest toilers
Find incombustible in boilers
Or for settling bills or monies due
To H.M. Inland Revenue.
His colleague Humphrey, sad to say,
Did nothing to enhance his pay
Or prestige, filling in his tome
With noughts-and-crosses, ribald rhyme,
Electronic timing for his Mini
Or amplifiers for home cine, With evenings spent in loosening hip-joints
In Soho's less salubrious clip joints. But Phil, who scorned such carnal larks As making love to birds in parks, Reorientated his intention
And sought his fortune by invention.
He patented in quick succession Brain-children in a long procession Including a unique device
For electronically catching mice.
His grateful Company, while pocketing
Royalties which sent the shares a-rocketing,
Displayed their human circumspection By raising Phil to Chief of Section And - mark the carrot, gentle reader Hinted a future as Group Leader; A circumstance which came to pass Helped by a calculating lass
Named Alpha Kerve, a lab assistant With ash-blonde hair and aim persistent Towards a matrimonial life
As a Lab. Manager's gracious wife A laudable ambition quite
For which she laboured day and night.
But colleague Humphrey stayed immune From Cupid's dart and, like a loon, Withdrew his hem from toil and strife In his laboratory life -
Conduct which labelled him "also ran" With his employers, Gargantuan, Who showed their grievous discontent By freezing Humph's emolument. Thus, on the ladder's lowest rung Humphrey precariously hung While Phil continued still to thrive The busiest bee in all the hive (not a foot wrong and ne'er a cropper A perfect ant to Humph's grasshopper

In you'll forgive this rhymster for The mixing of a metaphor.) And so the youthful years flew by With Phil advancing annually While Humphrey, wallowing in sloth, Sowed oats sufficient for them both An object-lesson and reminder
To keep our hooters to the grinder If in life's rat-race we'd succeed And thus aspire the field to lead.

\section*{EPILOGUE}

A figure in moth-eaten rags
Squats miserably upon the flags Outside the tube at Baron's Court "WIFE AND TV TO SUPPORT" Proclaims the placard on his chest And, touched by this oblique request, The passer-by donates his mite Despite the cost of living's bite; Pfennig and centimes tinkle in The thoughtfully provided tin. But stay! A sleek Rolls-Royce approaches And o'er the double line encroaches While from the car steps a retainer Depositing tuppence in the container And then his Christian duty done, Lord Humphrey Naild is driven on While Phil the coinage quickly clutches And gratefully his forelock touches. . . .

Patience, reader, I'll explain
'Til truth's as clear as windowpane Our Phil's inventive wells ran dry And brought him to redundancy, While Humphrey, tired of bread and water Craftily wed the Chairman's daughter Becoming, through this master-plan, The whizz-kid of Gargantuan.

\section*{MORAL:}

There isn't one.

\title{
mavis
}

ELECTRONIC CROSS-OVER


The Mavis 3 way electronic cross-over is intended for use primarily with music and speech amplifying systems. It enables the bass range, mid-range and treble range to be separately controlled. The cross-over frequency for each range can be specified if required but will be, in the standard unit, as follows:
Bass roll-off 45 c.p.s.
Bass to mid-crossing point 800 c.p.s.
Mid to treble crossing point 5000 c.p.s.
The unit's output is balanced 600 ohm Line for each chanmel capable of driving six 600 ohm balance sources. The input to the cross-over is also 600 ohm balance.

GENERAL SPECIFICATION

Size
Weight
Input
Output
Power Requirements
Optional extra
PRICE - £500
\(19^{\prime \prime} \times 12^{\prime \prime}\) deep \(\times 7^{\prime \prime}\) high (standard \(19^{\prime \prime}\) racking)
35lb.
0 dbm 600 ohm balance
+10 dbm 600 ohm balance
\(110 / 230\) volts \(50 / 60\) c.p.s. at 80 watts
approx.
Sub plate

WW-111 FOR FURTHER DETAILS

\section*{INTRODUCING THE P.A.S. 30/30}

PORTABLE MIXER


This mixer has been designed for mobile use in conjunction with high quality audio systems. It has basically 15 fully equalized input channels, plus 2 high level auxiliary input channels. The mixer can be used in two configurations, either 4 track full range output or 2 track output split into 3 channels each track, each channel controlled by an electronic cross-over. The remaining 2 tracks can be used either as full range tracks or re-mixed into tracks \(1 \& 2\) as sub-mixers The mixer also has 2 fully equalized independent monitor outputs and drive facilities pendent monitor outputs and drive facilities
for an external echo system. There is also an for an external echo system. There is also an
output for use with headphones to listen through for cueing each channel.

GENERALSPECIFICATION
Weight
Power Consumption
Input Impedince
Input level 75 modul Input level 75 modules Output level
Cue output leve
Equalisation range

Overall noise
Channel separation
\(38^{\prime \prime} \times 27^{\prime \prime} \times 12^{\prime \prime}\)
1901 b approximately 80 watts approximately
000 hm
600 ohm balanced -60 dbm
-0 dbm
\(-0 \mathrm{dbm}\)
+10 dbm all channels
- 300 milliwatts
\(\pm 14 \mathrm{db}\) treble
\(\pm 20 \mathrm{db}\) mid
\(\pm 14 \mathrm{db}\) bass
\(\pm 20 \mathrm{db}\) bass peak
better than - 60 db below full output
better than - 80 dbm

PRICE - E6,000 including freight case

WW—112 FOR FURTHER DETALS


\section*{P.A.S. 30/30}

This 30 Channel Desk is a development of the Mavis Four Group 15 Channel Mixer to meet the growing demands of modern P.A. and Studio work:' It is designed such that every channel may be operated with total flexibility in a four channel quadraphonic setup. and for purposes of live recording it is unique in the fact that a multi-track tape machine of up to 30 tracks may be directly coupled to the channels and a 4 track Tape Machine to the mains groups. The Mixer can then at a later stage be used for mixing down to a stereo or
quadmaster using the main group outputs. quadmaster using the main group outputs.
As a compromise between a P.A. Mixer and a conventional Studio
Desk, it differs from the latter in the fact that apart from the usual Desk, it differs from the latter in the fact that apart from the usual four main groups are employed when the desk is used in total; the line drives for recording are derived directly from each channel, and are fully equipped for patching in auxiliary equipment, and may be switched before or after the channels' "EQ" section.
The desk is built in three sections. Two wings (which may be used independently in stereo for P.A.) are equipped with fifteen channels each and a complete output arrangement including four groups and a stereo cross-over. The third section - the routing for the two wings and awn into four or two track. This is dit with quad control and mix down into four or two track. This is dealt with in Section B of the In-
struction Manual. Using an extra stereo cross-over each wing can drive a quadraphonic P.A. system.

\section*{GENERALSPECIFICATION}

The \(30 / 30\) Mixer is divided into four parts. A Centre Dask containOscillator and Master Quad and Par Oscillator and Master Quad and Pan facilities with 4 Master Faders,
There also can be built-1n remote control facilities for Dolby's Machine Control and Auto Tape Lecators. The Cacilties for Dobys Machine puts. 4 machine inputs. two foldback outputs and 4 monitor outputs also group break "in and out" facilities. There are also sockets to connect this desk to the two wings and a plug for the power supply. Two input wings which are mirror images. and contain 15 input

\section*{11a SHARPLESHALL ST., LONDON, N.W. 1}

Tel. 01-7227161/2/3/4
Telex: London 27655
modules, which have input trim and equalisation, also facilities which enable the module to supply a line level drive for a tape machine with n\% without equalisation also 4 group outputs which may be combined by switches to be used as quadraphonic output or a pan output.
There are facilities for 2 monitor or effects outputs and one echo output. The module has a switch which controls the output to group. off or cue.
There is also a switch which enables a break socket on the rear panel for effects drive and inDuts to be switched in and out
The fourth unit is the power supply which powers the Centre Desk and two wings and provides a 48 volts Phantom Microphone supply to the thinty microphone inputs.

Weioht Wina \(\quad 120 \mathrm{Kg}\) approx.
Centre 100 Kg approx
Power Consumption
Inout Impedance
Output Impedance
Maximum Input Sensitivit
Maximum Input Sensi
Microphone Input
Machine Input
Nominal Output Cue Output
Monitor Output Foldback Output Echo Output

100 Kg approx
500 watts
500 watts.
600 or 1200 ohms. Balanced
600 ohms. Balanced
\(-60 \mathrm{dbm}\)
0 dbm
+10 dbm PA
O dbm Machine
300 milliwatts
\(+10 \mathrm{dbm}\)
+10 dbm
+10 dbm

\section*{Sinclair Project 60}

\section*{New performance standards} ...new safety margins

Such are the results of using a PZ8 Mk. 3 to drive two 2.50 Mk .2 power amplifiers. Developed from the original Z.50, the MK. 2 has improved thermal stability. better regulated D.C. limiting to ensure more symmetrical output voltage swing with still less distortion at lower outputs and automatic transient overload protection. The PZ. 8 Mk .3 is the most advanced power supply unit ever to be made at a reasonable price. It cannot be damaged by direct power supply unit ever through overloading. because of an ingenous re-entrant current limiting principle used usually only in expensive laboratory equipment. Because output voltage is variable, the PZ8 Mk. 3 makes a worthwhile alternative where PZ.5 and PZ. 6 are vorage is variable, project 60 applications, particulariy since this most powerful of all Sinclair supply units can be operated from a smaller mains transformer. Together, the Sinclair supply units can be m. 20 Mk PZ8 Mk. 3 provide new standards of performance and reliability and these modules are compatible with earlier types in the Project 60 range.
Z.50 Mk. 2 SPECIFICATIONS

Input impedance \(100 \mathrm{~K} \Omega\)
Input (for 30 w into \(8 \Omega\) ) 400 mV
Signal to noise ratio, referred to fult \(0 / \mathrm{p}\) at 30 vHT 80 dB or better Distortion \(0.02 \%\) up to 20 W at \(8 \Omega\) See published curve
Frequency response 10 Hz to more than \(200 \mathrm{KHz} \pm 1 \mathrm{~dB}\)
Max. supply voltage 45 V ( \(4 \Omega\) to \(8 \Omega\)
speakers) ( \(50 \mathrm{v} 15 \Omega\) speakers only)

Min. supply voltage \(9 v\) Load impedance - minimum: \(4 \Omega\) at \(45 v\) HT
Load impedance - maximum: safe on open circuit

\section*{\(\mathbf{6} 5.48+\) V.A.T}

PZ. 8 MK. 3 SPECIFICATIONS Nominal working output 45 V . Adjustable between \(20 \& 50 \mathrm{~V}\).
\(£ 7.98{ }_{79 \mathrm{p}}^{\mathrm{VAT}}\)
Mains Transformer \(£ 5.98+\) VA.T.59p


\section*{Other power supplies}

In addition to the remarkable Sinclair PZ.8 Mk.III as described. there are two other power units available, which should be chosen according to their types in order to buy to best advantage. All are for operation from A.C. mains 240 V .
PZ. 530 volt, unstabilised
PZ. 635 volt, stabilised (Not suitabie for Super
IC.12).
\(£ 7.98\)
+ V.A.T. \(79 p\)

\section*{Guarantee}

If, within 3 monthes of purchasing any product direct from Sinclair Radionics Ltd., vou are dissatisfied with it, your money will be refunded at once. Many Sinclair appointed Stockists also offer this same guarantee in co-operation with Sinclair Radionics Lid
Each Project 60 module is lested before leaving our factory and guaranteed to work perfectly. Should any defect arise in to you. A small charge may be made in those cases where damage arises through miss-use. No charge is made for postage by surface mail. Air Mall charged at cost.

\section*{Typical Project 60 applications}
\begin{tabular}{|c|c|c|c|}
\hline System & The Units to use & together with & Units cost \\
\hline Simple battery record player & 2.50 & Crystal P.U., 12 V battery volume control, etc. & \[
\begin{aligned}
& £ 5.48 \\
& + \text { V.A.T. } 54 \text { p }
\end{aligned}
\] \\
\hline Mains powered record player & Z.50, PZ. 5 & Crystal or ceramic P.U. volume control, etc. & \[
\begin{array}{r}
\text { £10.46 } \\
+\quad \text { V.A.T. £1.04 }
\end{array}
\] \\
\hline 12 W. RMS continuous sine wave stereo amp. for average needs & \[
\begin{aligned}
& 2 \times 2.50 . \text { Stereo } \\
& 60 ; \text { PZ.5 }
\end{aligned}
\] & Crystal, ceramic or mag. P.U., F.M. Tuner, etc. & \[
\begin{aligned}
& £ 25.92 \\
& +V . A T . \\
& £ 2.59
\end{aligned}
\] \\
\hline 25 W . RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers & \[
\begin{aligned}
& 2 \times \text { Z.50. Stereo } \\
& 60 ; \text { PZ. } 6
\end{aligned}
\] & High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc. & \[
\begin{aligned}
& \text { £28.92 } \\
& \text { £ VA. } .89
\end{aligned}
\] \\
\hline 80W. (3 0hms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms) & \begin{tabular}{l}
\(2 \times \mathrm{Z} .50 \mathrm{Mk} .2\). \\
Stereo 60 ; PZ. 8 \\
Mk. 3 transformer
\end{tabular} & As above & \[
\begin{aligned}
& \text { £34.90 } \\
& + \text { V.A.T. }
\end{aligned}
\] \\
\hline Indoor P.A. & Z.50 Mk.2. PZ. 8 Mk. 3 transformer & Mic., guitar, speakers, etc., controls & \[
\begin{aligned}
& £ 19.44 \\
& +\quad V \text { A.T.£1.94 } \\
& \hline
\end{aligned}
\] \\
\hline
\end{tabular}

\footnotetext{
A.F.U. (f.5.98 + V.A.T. 59 p) may be added as required.
}

\section*{the world's most advanced high fidelity modules}

\section*{Q. 16 high fidelity loudspeaker}

The 016 employs original and by now well proven acoustic principles in which a special driver assembly is meticulously matched to a uniquely designed cabinet. In performance it comfortably stands comparison with very much more expensive loudspeakers. A, solid teak surround is used with a special all-over cellular black foam front chosen boin for its appearance and ability to pass all audio frequencies without masking.
Specifications
Construction: A sealed seamless sound or pressure chamber is used with in ternal baffle, and special high fux driver
Loading: Up to 14 watts RMS, into 80 hms
Frequency response: From 60 to \(16,000 \mathrm{~Hz}\)
Size and styling: 248 mm square \(\times 120 \mathrm{~mm}\) deep
( \(9 \frac{3}{4}\) " \(\times 4 \frac{3}{4}\) ") with neat pedestal base



\section*{Project 605}

\author{
\section*{the} simple \\ way to build a Project 60 system without soldering
}

For the many audio enthusiasts anxious to build to high standards without too many involvements, there could be nothing better or simpler than Project 605 it offers the advantages of Project 60 and is absolutely complete down to the last plece of wire cut to length. Whilst not as powerful as assemblies using Z.50 power amplifiers, we know from experience that there are many for whom the specifications of Project 605 are ideal, particularly in relation to the environment in which it is required to be used. In Project 605 you have everything necessary 10 build a versatile Project 60 thirty watt high fidelity amplifier system suitable for all domestic requirements. The conventent pack includes two Z. 30 power amplifiers, a Stereo 60 pre-amp control unit and the special Masterlink unit to and from which all input and output connections are made. For power a PZ. 5 is provided. Building is parıcularly easy since all necessary leads are supplied colour coded, cut to length and terminated by contact clips which connect firmly to the modules. There is absoiltely no soldering to be done. Complete with comprehensive, easy to follow instructions manual

\section*{£29.95}
V.A.T. £2. 99

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Please send leaflet and name and address of my nearest Sinclair stockist

Name

Address

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\section*{Lanys U.K's LaRGEST RANGE OF BRANDED AND GUARANTEED DEV/CES. (Quantity Discounts 10\% \(12+\), 15\% 25 + , 20\% \(100+\) ) (Any one type except where quantify discounts show) Min. Order \(£ 1.00\) please. Post 10 p.}


\title{
Lenrys U.K's LaRgest range of electronic components AND EQUIPMENT AT BARGAIN PRICES \\ Latest Catalogue price 55p post paid. Complete with Discount Vouchers
}


404-406 Electronic Components and Equipment 01-4028381* -354-356 High Fidelity and Tape Equipment 01-402 5854/4736 309 PA-Disco-Lighting High Power Sound 01-723 6963 303 Special offers and bargains store


\section*{SEMICONDUCTORS}
\begin{tabular}{|c|c|c|c|}
\hline 2N699 & 0.25 & BC184L & 0.11 \\
\hline 2N1613 & 0.20 & BC212L & 0.12 \\
\hline N1711 & 0.25 & BC214L & 0.14 \\
\hline N2926G & 0.10 & BCY72 & 0.13 \\
\hline 2N3053 & 0.15 & BF257 & 0. \\
\hline 2N3055 & 0.45 & BF259 & 0.47 \\
\hline 2N3442 & 1.20 & BFR39 & 0.25 \\
\hline 2N3702 & 0.11 & BFR79 & 0.25 \\
\hline N3703 & 0.10 & BFY50 & 0.2 \\
\hline 2N3704 & 0.10 & BFY51 & 0.20 \\
\hline 2N3705 & 0.10 & BFY52 & 0.20 \\
\hline 2N3706 & 0.09 & M3481 & 1.20
1.30 \\
\hline 2N3707 & 0.10 & MJE521 & 0.60 \\
\hline 2N3708 & 0.07 & MPSA05 & 0.3 \\
\hline 2N3709 & 0.09 & MPSAI 2 & 0.5 \\
\hline 2N3710 & 0.09 & MPSA14 & 0.35 \\
\hline 2N3711 & 0.09 & MPSA5 & 0.35 \\
\hline 2N3819 & 0.23 & MPSA65 & \\
\hline 2N3904 & 0.17 & MPSUOS & 0.60 \\
\hline 2N3906 & 0.20 & MPSU55 & 0.70 \\
\hline 2N4058 & 0.12 & SN72741P & \\
\hline 2N4062 & 0.11 & SN72748 & 0.58 \\
\hline 2N4302 & 0.60 & THBI & +10 \\
\hline 2N5087 & 0.42 & TIP30A & 0.60 \\
\hline 2N5210 & 0.54 & TIP31A & 0.60 \\
\hline 2N5457 & 0.30 & TIP32A & 0.70 \\
\hline 2N5830 & 0.30 & TIP33A & 1.00 \\
\hline 40361 & 0.40 & TIP34A & 1.50 \\
\hline 40362 & 0.45 & TTP4.a & 0.74
0.90 \\
\hline BC107 & 0.08 & TIP3055 & 0.60 \\
\hline BC108 & 0.08 & 1808720 & 0.50 \\
\hline BC109 & 0.08 & IB40K2 & - \\
\hline BC125 & 0.15 & IN914 & 0.07 \\
\hline BCl26 & 0.15 & \({ }_{1544}\) & 0.07 \\
\hline BC182K & 0.10 & 15920 & 0.10 \\
\hline BC212K & 0.12 & 153062 & 0.25 \\
\hline BC182L & 0.10 & 5805 & 1.20 \\
\hline
\end{tabular}

\title{
HI-FI NEWS 75 WATT AMPLIFIER BY J. L. LINSLEY-HOOD
}

\author{
Published Nov. 1972 to Feb. 1973
}

DESIGNER APPROVED KIT


SLIMLINE STYLE CHASSIS DIMENSIONS: \(17.0 \mathrm{in} . \times 2.0 \mathrm{in} . \times 12.0 \mathrm{in}\) This slimline unit has been made practical by the use of a specially designed TOROIDAL TRANSFORMER and highly compact printed circuit boards which have been fully tested and approved by Mr. Linsley-Hood.

\section*{FREE TEAK CASE}

WITH 75 WATT PER CHANNEL COMPLETE AMPLIFIER KITS

Total cost of individually purchased packs:
£63.95
Cost of complete kit: £56.60
TRADE ENQUIRIES WELCOME
P.S. Full circuit description in handbook

30p

\section*{FOR FURTHER DETAILS PLEASE WRITE TO:}
75 WATTS PER CHANNEL
BANDWIDTH (3dB) \(3 \mathrm{HZ}-40 \mathrm{KHZ}\) DISTORTION LESS THAN 0.0I \%UNCONDITIONAL STABILITYCOMPONENT PACKS
Pack
I Fibre glass printed circuit board for power amp ..... \(£ 0.75\)
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3 Set of semi-conductors for power amp. (highest voltage version) ..... \(£ 5.50\)
4 Pair of 2 drilled, finned heat sinks ..... 10.80
5 Fibre glass printed circuit board for pre-amp. ..... 61.10
6 Set of low noise resistors, capacitors, pre-sets for pre-amp ..... \(\varepsilon 2.70\)
7 Set of low noise, high gain semi-conductors for pre-amp ..... £2.10
8 Set of potentiometers (including mains switch) ..... € 1.55
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10 Toroidal transformer complete with magnetic screen/housing primary: 0-117-234 V. secondaries: 33-0-33 V.24-0-24 V., electrostatic screen\(£ 9.15\)
II Fibre glass printed circuit board for power supply ..... 60.55
12 Set of resistors, capacitors, secondary fuses, semi conductors for power supply ..... \(€ 3.50\)
13 Set of miscellaneous parts including DIN skts., mains input skt. fuse holder, interconnecting cable, contro knobs ..... £3.25
14 Set of metal workparts including silk screen printed fascia panel and all brackets, fixing parts, etc. ..... C6.30
15 Handbook, based on Hi-Fi News articles ..... 60.30
16 Teak cabinet ..... c7.352 each of packs I-7 inclusive are required for completestereo system. \\ PORTWAY INDUSTRIAL ESTATE, ANDOVER : HANTS \\ MAIL ORDER ONLY POST FREE TO U.K. OVERSEAS AT COST \\ U.K. Orders Subject to 10\% V.A.T. Surcharge \\ \title{
\section*{POWERTRAN ELECTRONICS}
} \\ \title{
\section*{POWERTRAN ELECTRONICS}
}

\section*{Basic Component Set}

Set of semi-conductors, resistors, capacitors, printed circuit boards for stereo power amp, pre-amp. and power supply. \(\pm 31.35\)

Handbook Included


MARCONI SIGNAL GENERATOR TYPE TF-144G: Freq. \(85 \mathrm{Kc} / \mathrm{s}-25 \mathrm{Mc} / \mathrm{s}\) in 8 ranges. Incremental: \(\pm 1 \%\) at \(1 \mathrm{Mc} / \mathrm{s}\). Output: continuously variable 1 micro volt to 1 volt. Output Impedance: 1 microvolt to 100 millivolts, 10 ohms \(100 \mathrm{mV}-1\) volt - 52.5 ohms. Internal Modulation: \(400 \mathrm{c} / \mathrm{s}\) sinewave \(75 \%\) depth. Externa
Modulation: Direct or via internal amplifier. A.C. mains \(200 / 250 \mathrm{~V}, 40-100 \mathrm{c} / \mathrm{s}\) Consumption approx. 40 watts. Measurements \(29 \times 121 \times 10 \mathrm{in}\). Secondhand condition. \(£ 27 \cdot 50\) each, Carr. \(£ 1 \cdot 50\).
T. 1509 TRANSMITTERS (FOR EXPORT ONLY): General-purpose HF communications transmitter for use in fixed or mobile ground stations. Hand or high-speed keying. Crystal or MO control, w. Fircuit.CW, MCW and R/T. Frequency: 15 to \(20 \mathrm{Mc} / \mathrm{s}\). Modulation: \(100 \%\) \(0 /\) put impedance: 50 ohms. Audio input: 600 ohms . Valves : Power Amplifier \(2 \times 813\) and Modulator \(2 \times 813\). Power requirements \(200-250\) volts a.c. 50 cycles. Power out put 300 watts. Dimensions 2 ft . 6 in . W. \(\times 2 \mathrm{ft}\). D. \(\times\) 5 ft . H. Weight : 800 lbs . Excellent condition, price \(£ 225.00\) each.
AN/ARC-27 TRANSMITTER/RECEIVER (FOR EXPORT ONLY): Frequency \(225-400 \mathrm{mc}\). 1750 channels 100 Kc apart with 18 preset channels Modulation: am. Power output 9 watts. Receiver is superheterodyne. Max. output 2 watts. Antenna: 50 ohm impedance. Power requirements 24 v d.c chone. Price \(£ 250.00\) each secondhand, excellent condition phone. Price \(£ 250.00\) each secondhand, excellent condition.
nput. 24 v d.c. output @ 41 amps fully smoothed. \(£ 45.00\) each 250 volts a.c
FREQUENCY METER BC-221: \(125-20,000 \mathrm{Kc} / \mathrm{s}\), complete with original calibration charts. Checked out, working order. \(£ 18 \cdot 50+£ 1.00\) carr. BC-221 E2-00 carr.

CT. 52 MINIATURE OSCILLOSCOPE: Portable. Operates from 115 V or \(250 \mathrm{~V} 50-60 \mathrm{c} / \mathrm{s}\); or \(180 \mathrm{~V} 500 \mathrm{c} / \mathrm{s}\). A small compact tropicalised Enstrument designed to meet requirements of radar and communication engineers and general electronic service. Measures \(9 \mathrm{in}, \times 8 \mathrm{in}\). \(\times 6 \frac{\mathrm{hin}}{} \mathrm{in}\). Time base \(10 \mathrm{c} / \mathrm{s}-\) amplifier up to 38 dB gain. Bandwidth up to \(1 \mathrm{Mc} / \mathrm{s}\). Single sweer facilities.


TUNING UNIT: 24V geared motor driving double 25pf double spaced variable Capacitor. One m/c relay and 2 other relays. \(£ 2 \cdot 50\) each 30 p post, gocid condition. C42, 2C46, 1B40 (complete with associated capacitors and screening) 3 alves counters \(0-999\). Valves 6 AL5 and \(8 \times 6 \mathrm{AK} 5\). \(£ 10.00\) plus 60 p post, good condition.
MODULATOR UNIT: complete with transformer and \(2 \times 807\) valves mounted in 19 in. chassis \(\times 8\) in. high \(\times 8\) in. deep. \(£ 4 \cdot 50\) secondhand cond., or \(£ 6.50\) new cond. Carriage \(£ 1\)
RF UNIT: suitable for use with the above unit. Complete with \(2 \times 3 \mathrm{E} 29\) valves Ideal for conversion to 4 metres. \(£ 5\) secondhand cond., or \(£ 7.50\) new cond.
POWER SUPPLY UNIT PN-12A: 230V a.c. input 50-60 c/s. 513V and 1025V @ 20 mA output. With 2 smoothing chokes \(9 \mathrm{H}, 2\) Capacitors, 10 Mfd 1500 V and 10 Mfd 600 V . Filament Transformer 230 V a.c. input. 4 Rectifying Valyes type 5 Z 3 \(\times 5 \mathrm{~V}\) windings @3 Amps each, and 5V@6 Amp and 4V @ 0.25 Amp . Mounted on steel base \(19^{\prime \prime} \mathrm{Wxl} 1^{\prime \prime} \mathrm{Hx} 14^{\prime \prime} \mathrm{D}\). (All connections at the rear.) Excellent condition

AUTO TRANSFORMER: \(230-115 \mathrm{~V}, 50-60 \mathrm{c} / \mathrm{s}, 1000\) watts, mounted in a strong steel case \(5^{\prime \prime} \times 6 \frac{1}{2 \prime}^{\prime \prime} \times 7^{\prime \prime}\). Bitumen impregnated. \(£ 7\) each, Carr. 75 p . \(230-115 \mathrm{~V}\) \(50-60 \mathrm{c} / \mathrm{s}, 500\) watts. \(7^{\prime \prime} \times 5^{\prime \prime} \times 5^{\prime \prime}\). Mounted in steel ventilated case. \(£ 4.00\) each,
Carr. 75 p .
MODULATOR UNIT: 50 watt, part of BC-640, complete with \(2 \times 811\) valves, microphone and modulator transformers etc. \(£ 7 \mathbf{5 0}\) each, 75 p carr.
CATHODE RAY TUBE UNIT: With 3in. tube, Type 3EG1 (CV1526) colou green, medium persistence complete with nu-metal screen, APN-1 INDICATOR METER, \(270^{\circ}\) Movement. Ideal for making rev. counter
\&1.25, post 30 p.
AIRCRAFT SOLENOID UNIT S.P.S.T.: \(24 \mathrm{~V}, 200 \mathrm{Amps}\), \(\mathbf{£ 2}\) each, \(\mathbf{3 0} \mathrm{p}\) post. DECADE RESISTOR SWITCH: 0.1 ohm per step. 10 positions. 3 Gang, each 0.9 ohms. Tolerance \(\neq 1 \%\) £ 3 each, 25 p post. 90 ohms per step. 10 positions
total value 900 ohms. 3 Gang. Tolerance \(+1 \%\). 3.50 each TF-1041B
TF-1041B VALVE VOLTMETER: Measures 25 mV to \(300 \mathrm{~V}, 20 \mathrm{c} / \mathrm{s}\) to 1500 \(\mathrm{Mc} / \mathrm{s}\) a.c. Also 10 mV to 1000 V d.c. Resistance 0.02 ohms to 500 Meg . oims. Powe requirements \(200-250\) volts a.c. Secondhand, excellent con. \(£ 35 \cdot 00\). Carr. £. 1.
VARIAC TRANSFORMERS: Input 115 V , output \(0-135 \mathrm{~V}\) at 2 Amps. \&3 each
RACK CABINETS: (totally enclosed) for Std. 19 in . Panels. Size 6 ft . high \(\times 21\) in. Wide \(\times 16\) in. deep, with rear door. \(£ 12\) each, \(£ 2.50\) Carr. OR 4 ft . high \(\times 23\) in. wide \(\times 19 \mathrm{in}\). deep, with rear door. \(\mathbf{£ 8} 50\), each, £2 Carr.
INSTRUMENT CABINETS: \(19^{\prime \prime} \mathrm{W} . \times 16^{\prime \prime} \mathrm{H} . \times 16^{\prime \prime} \mathrm{D} . \quad £ 500+£ 1.25 \mathrm{carr}\)
\(19^{\prime \prime} \mathrm{W} . \times 10^{\prime \prime} \mathrm{D} . \times 5^{\prime \prime} \mathrm{H} . £ 2.50+6\).
FUEL. INDICATOR Type 113R: 24V complete with 2 magnetic counters \(0-9999\), with locking and reset controls mounted in 3in. diameter case. Price £2
each, 30 p post. each, 30p post.
TS-418/URM49 SIGNAL GENERATOR: Covers \(400-1000 \mathrm{MHz}\);ange. CW Pulse or AM emission. Power Range \(0-120 \mathrm{dbm} . £ 125\) each. Carr, £1.50.
ALL U.K. ORDERS SUBJECT TO 10\% VALUE ADDED TAX.

TN/130/APR. 9 UHF TUNING UNIT: Freq. \(4300-7350 \mathrm{MHz}\). IF Output 160 MHz with bandwidth of 20 MHz and is electrically tuned by a d.c. reversible notor. \(227 \cdot 50\) each. Carr. £1.
APR-4 AM RADIO RECEIVER: \(90-1000 \mathrm{MHz}\). This receiver is suitable for monitoring and measuring trequencies as well as relative signal strength. Power Supply \(115 \mathrm{~V} 50 \mathrm{c} / \mathrm{s}\). £ 100 each. Carr. £2.
SIGNAL GENERATOR TS-497B/URR: (Boonton). Freq. 2-400 Mc/s in bands. Internal Mod. 400 or \(1000 \mathrm{c} / \mathrm{s}\) per sec. External Mod. 50 to \(10,000 \mathrm{c} / \mathrm{s}\) o/put Voltage \(0 \cdot 1-100,000\) microvolts -30 for sine wave. Am or Pulse Carrier \(685 \mathrm{each}+£ 1.50 \mathrm{carr}\).
CLASS "D" WAVEMETER NO. 1 MK. II: Crystal controlled heterodyne requency meter covering \(2-8 \mathrm{MHz}\). Power supply 6 V d.c. Good secondhand cond.
7.50 each. Post 60 p . \({ }_{6} 750\) each. Post 60p
RCA TE-149 HETERODYNE WAVEMETER: V-cut, 1 MHz crystal ( \(0.005 \%\) ) ccuracy better than \(0.02 \%\). Dial directly calibrated every 1 KHz from 2.5 .5 MHz Useful harmonics up to 20 MHz . Provision for fitting internal dry batteries. "As new" complete with Manual and Spares. £14 each. Carr. 75p.
POWER UNIT TYPE 24: (for R. 216 Receiver) A.C. operated \(100-125 \mathrm{~V}\) or 200-250V, 50c/s. "As new" \(£ 10\) each. Carr. 75p.
ROTARY INVERTERS: TYPE PE.218E-input 24-28V d.c., 80 Amp 4,800 rpm. Output 115 V a.c. \(13 \mathrm{Amp} 400 \mathrm{c} / \mathrm{s}\). 1 Ph . P.F.9. \(£ 17 \cdot 50\) each. Carr. \(£ 1 \cdot 50\). POWER SUPPLY: 230V a.c. input; 3000V @ \(2 \cdot 5 \mathrm{~mA}\); 4v@1 Amp, 300-0-300 00mA; 6V @ 7 Amp ; 6V @ 3 Amp . With smoothing capacitors etc. \(\mathbf{£ 1 0 \cdot 0 0}\) each £1-50 carr.

ACTUATOR UNIT: With 115 V d.c. geared motor; o/put 12.5 rpm ; torque 16 ins. oz; reversible; microswitches and potentiometer. \(£ 3.50\) ea. +40 p post. DALMOTORS: \(24-28 \mathrm{~V}\) d.c. at 45 Amps , 750 watts (approx. 1 hp ) 12,000rpm £5 each, 60p post
MOTOR: 240 V single phase, \(2,400 \mathrm{rpm} .1 / 40 \mathrm{H} . \mathrm{P}\). approx. Price \(£ 1 \cdot 75\) each, 30p post.

CONDENSERS: 30 mfd 600 V wkg. d.c., \(£ 3.50\) each, post 50 p .10 mfd 1000 V Wkg. 80 p , post 30 p .8 mfd 2500 v £5, carr. 80 p .8 mfd 600 v 45 p , post 15 p .8 mfd \(1_{0}^{\circ} 300 \mathrm{v}\) d.c., \(£ 1 \cdot 25\), post 25 p .4 mfd 3000 v wkg . \(£ 3\), post \(50 \mathrm{p} .4 \mathrm{mfd} 2000 \mathrm{v} £ 2\), post \(40 \mathrm{p} .4 \mathrm{mfd} 600 \mathrm{v}, 2\) for \(£ 1.00\), post 30 p . Capacitor \(0.125 \mathrm{mfd} 27,000 \mathrm{v}\) w.kg \(£ 3.75\), post 50 p .225 mfd 25 Kv wkg. £20, carr. . 3.2 mfd 12.5 Kv wkg. TCC RL
\(7002-97, ~ £ 8 \cdot 50\), carr. \(£ 1.10 \mathrm{mfd} 3 \mathrm{Kv}\) wkg, \(55^{\circ} \mathrm{C}\). TCC oil filled, \(£ 7 \cdot 50\), carr. \(£ 1\). \(7002-97, £ 8 \cdot 50\), carr. \(£ 1.10 \mathrm{mfd} 3 \mathrm{Kv}\) wkg, \(55^{\circ} \mathrm{C}\). TCC oil filled, \(£ 7 \cdot 50\), carr. \(£ 1\).
5 ml 1 mfd 3 Kv wkg. \(55^{\circ} \mathrm{C}\). \(£ 6 \cdot 50\), carr. \(£ 1.12 \mathrm{mfd} 1500 \mathrm{v}\) d.c. wkg. \(£ 3 \cdot 50\), post 50 p. CONTROL PANEL: 230 v. A.C., 24 v. D.C. @ \(2 \mathrm{amps}, ~ £ 2.50\) each, carr. 75 p . OHMITE VARIABLE RESISTOR: 5 ohms, \(5 \frac{1}{2} \mathrm{amps}\); or 40 ohms at 2.6 amps 500 ohms, 0.55 amps . Price (either type) \(£ 2\) each, 30 p post each.
TX DRIVER UNIT: Freq. 100-156 Mc/s. Valves \(3 \times 3\) C 24 's; complete with filament transformer 230 v. A.C. Mounted in 19 in . panel, \(£ 4.50\) each, carr. 75 p . AR88 RECEIVER: List of spares, 5 p.
TELEPRINTER EQUIPMENT, REPERFORATORS, READERS, and AUTO TRANSMITTERS ETC. Send for list, 5 p.
REDIFON TELEPRINTER RELAY UNIT NO. 12: ZA-41196 and powe supply \(200-250 \mathrm{~V}\) a.c. Polarised relay type 3 SEITR. \(80-0-80 \mathrm{~V} 25 \mathrm{~mA}\). Two stabicondition \(£ 7.50\), Carr. 75 p .
WESTON INDUSTRIAL THERMOMETER MODEL 221: \(0-100^{\circ} \mathrm{C}\). 3in dia. scale. Accuracy \(1 \%\). Precision made coil within-coil structure. Changes in temperature cause a rotary action of the Helix turning the shaft to which the pointer is mounted. \(£ 2 \cdot 80\) each 30 p post. Unused condition
TRANSMITTER UNITS: Complete with 12 V vibrator unit QQVO3-20A and 5 other valves with modulation transformer, etc. Two crystal controlled channels. Suitable for conversion to 2 metres. \(£ 5+£ 1\) carr.
THERMOCOUPLE METER: Scale 3.5 AE 2 in . square flush mounting \({ }^{2} 2.50+25\) p post
TS 15C/AP FLUXMETER: Used to provide qualitative measurements of flux densities between pole faces of magnets. Range \(1200-9600\) gausses. \(\pm 2 \% . \mathrm{S} / \mathrm{hand}\)
good cond. \(£ 25+60 \mathrm{p}\) post. good cond. \(\mathbf{t 2 5}+60 \mathrm{p}\) post.
SYNCHRO DISTORTION AND MARGIN TEST SET: (Onwood Type 4A2) S/hand excellent cond. \(£ 85\) each. Carr. \(£ 2\).
MASTER SYNCHRO TEST SET T. 101031 (U.S.A.): 115 volts \(400 \mathrm{c} / \mathrm{s}\). S/hand cond. \(£ 15\) each \(+£ 1\) carr
MAGSLIP TESTER NO. 2 MK. I: S/hand cond. \(£ 25\) each \(+£ 1\) carr
SYNCHROS: and other special purpose motors available. Send for list. S.A.E PANORAMIC ADAPTOR TYPE ALA2: Suitable for use with APR-1, APR-4 and other Receivers having an I.F. frequency of 30 MHz . Will display signal up to 5 MHz either side of the received frequency. Power Supply 115 V a.c. \(400 \mathrm{c} / \mathrm{s}\) Tube 3PB1 with nu-metal screen. \(£ 8 \cdot 50\) each. \(£ 1\) carr. S/hand cond.
MUIRHEAD PAMETRADA WAVE ANALYSER D-489-D: Primarily used for the analysis of complex vibration waveforms, but will measure audio and power frequency waveforms from \(19 \mathrm{c} / \mathrm{s}\) to \(21 \mathrm{kc} / \mathrm{s}\). Complete with power supply unit 230 volts \(50 \mathrm{c} / \mathrm{s}\). S/hand good cond. \(£ 82 \cdot 50+£ 2 \mathrm{carr}\).
D-652 L.F. MODULATOR: Suitable for use with the above Wave Analyser D-489-D enabling the analysis of low frequencies between 2 and \(20 \mathrm{c} / \mathrm{s}\). S/hand good cond. \(£ 25\) each \(+£ 1\). carr.
AUTOMATIC VIBRATION EXCITER CONTROL UNIT TYPE 1016 Manufactured by Bruel \& Kjoer. \(5-5000 \mathrm{c} / \mathrm{s}\). per second. S/hand very good cond. \(£ 90+£ 2\) carr.
INSULATION TEST SETS: A.C. or D.C. \(0-5 \mathrm{kV} . £ 22.50\). S/hand cond AND 0-3 kV. Positive and negative outputs, fine and course control. \(£ 17 \cdot 50\) S/hand cond. Carr. both types \(f 2\).
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Regponae ONLY
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\(250-0-250 \mathrm{v}, 1\)
60 mA PULLY 8HROUDED UPRIGHT MOUSTM \(250-0-250 \mathrm{v} .60 \mathrm{~mA} ., 6.3 \mathrm{v} .2 \mathrm{a} .0-5-5-6.3 \mathrm{v} .2 \mathrm{~s}\).
\(250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6.5 \mathrm{v} .4 \mathrm{a} ., 0-\mathrm{s}-6.3 \mathrm{v} .3 \mathrm{a}\). \(3000-0-300 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a} ., 0-5-5.6 .3 \mathrm{v} .3 \mathrm{a}\).
 For Moliard 510 Ampiffer.

 \(450-0-450 \mathrm{v} .250 \mathrm{~mA} .6 .3 \mathrm{v} .4 \mathrm{a} ., \mathrm{c}\) e.t. 5 v .3 am TOP SEROODED DROP-THROUGH TYPE
 \(250-0.250 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a} ., 6.9 \mathrm{v} .1 \mathrm{a}\).
\(350-0-350 \mathrm{v} .80 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a} .0-5-6.3 \mathrm{v} .2 \mathrm{a}\)
 \(300-0.300 \mathrm{v}, 100 \mathrm{~mA}, 6.3 \mathrm{z}, 4 \mathrm{a}, 0,0-6-6.3 \mathrm{v} .3 \mathrm{a}\).
\(300-0.300 \mathrm{v}, 180 \mathrm{~mA}, 6.3 \mathrm{z}\)
 \(350-6.350 \mathrm{v}\). 100 mA ., 6.3 v . \(4 \mathrm{a} . .00-5-6.3 \mathrm{v}\).





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 OUTPUT TRAKSFORMERS Push-Pull 8 watts EL84 to 3 O or 15 @... or 150 ............................... Push-Pull ELS4 to 3 or 150 10-12 watte.
Push-Pull UItra Lnear for Mullard 510 , etc. Push-Pull \(15-18\) watte, sectionally wound \({ }^{6 L 6}\) ET66, etc., for 3 or 150. \(\qquad\) BATTERY MAINS BATTERY/MAINS CONVERSION UNITS
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where normal 200-200v. AD
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 sound quality in suitable enclosure.
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 MODEL \(803 \mathrm{~T} 8^{\prime \prime} 15 \mathrm{w}\). with parasitic Tweeter Response 25 Hz to 15 KEz . Gauss
\(13,000 \mathrm{Imp} 3\) or 8.15 ohms. ONLY
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 55 p
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1109
\] & \[
1010
\] & & 100 u \\
\hline C & 1/20W & 5\% & 82S2-220KS & E12 & 9 & \({ }_{3}\) & & \\
\hline c & \(1 / 8 \mathrm{~W}\) & 5\% & \(4 \cdot 7 \Omega-470 \mathrm{~K} \Omega\) & E24 & 1 & 0.9 & & 975ntt \\
\hline C & \(1 / 4 \mathrm{~W}\) & 5\% & 4.78-10M \(\Omega\) & E12 & 1 & 0.9 & & 0.75 nett \\
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\hline C & 1 W & 5\% & 4.7R-10M \(\Omega\) & E12 & 2.5 & 2 & & 16 nett \\
\hline MO & \(1 / 2 \mathrm{~W}\) & 2\% & 10S-1MS2 & E24 & 4 & 3 & & nett \\
\hline ww & 1W & 10\% \({ }^{\text {a }} 1 / 20\) 2 & 0.22S-3.9 & E12 & 7 & 7 & & \\
\hline WW & 3W & 5\% & 1 \(\mathrm{S}^{2-10 \mathrm{~K} \Omega}\) & E12 & 7 & 7 & \({ }^{6}\) & \\
\hline \multicolumn{9}{|l|}{\multirow[t]{2}{*}{\begin{tabular}{l}
Codes: \(\mathrm{C}=\) carbon film, high stability, low noise. \\
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\end{tabular}}} \\
\hline & & & & & & & & \\
\hline \multicolumn{9}{|l|}{Values: value and power rating. NOT} \\
\hline \multicolumn{9}{|l|}{\multirow[t]{2}{*}{E12 denotes series: \(10,12,15,18,22,27,33,39,47,56\), mixed values. 68, 82 and their decades. tions of one penn}} \\
\hline & & & & & & & & \\
\hline \multicolumn{9}{|l|}{E24 denotes series. as E12 plus 11, 13, 16, 20, 24, 30, 36, of resistor order.)} \\
\hline
\end{tabular}

\section*{TRANSISTORS BY SIEMENS AND NEWMARKET}

\author{
2N3055 npn silicon power \\ ACI 53 K pnp germanium low power ACI76K npn germanium low power ADI62 npn germanium medium power AF139 pnp germanium medium power AF19 pnp germanium BC107-13p; BC108-12p; BC109-13p BC167-11p; BC 168 -10p; BC 169 - 11 p BC257-12p; BC258-11p; BC179-22p Standard groupings available. \\ Very many other types listed, described and illustrated in catalogue
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CD. 1220 OSCILLOSCOPE, with dualtrace Plug-in, (CX1257) \(\mathrm{DC}-24 \mathrm{MHZ}\). \(\mathrm{f125}\). \((\mathrm{CX1256}) \mathrm{DC}-40 \mathrm{MHZ}, \mathrm{E} 25\).
Wide band Plug-in. SOLARTRON OSCILLATOR (CO546) \(25 \mathrm{~Hz}-500 \mathrm{KHz} £ 50\). OVERLOAO CUT.OUTS. Panel mounting \(\left(1 \frac{3}{4} \times 1 \frac{1}{6} \times \frac{1}{2} \mathrm{in}\right.\).)
BOO M/A/t. \(8 \mathrm{amp} / 10\) amp. 35 p ea. P.P. 5 p . BULK COMPONENT OFFER. Resistors/Capacitors. All
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\(500 \mathrm{M} / \mathrm{A} .-80 \mathrm{~V} @ 500 \mathrm{M} / \mathrm{A}\left(9 \times 6 \times 5 \frac{1}{2} \mathrm{in}\right.\).) New. \(£ 8.50\) with
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\section*{TRANSFORMERS}

Ler. Tim
250 WATT ISOLATION TRANSFORMER. 240 v . double wound E3.25. P.P. 50 p .
E.H.T. TRANSFORMER. Prim. 240 v . Sec. \(2 \cdot 5-0-2.5 \mathrm{kV}\). 12 mA .; 7.5 v .1 amp. 2.5 v .2 amp. \(£ 2.50\). P.P. 25 p
E.H.T. TRANSFORMER. Prim. 240v, Sec \(1800 \mathrm{v}, 50 \mathrm{~mA}\).
L.T. TRANSFORMER. (Shrouded) Prim. 200/250v.

Sec. 20/40/60v. 2 amp. \(£ 2\) ea. P.P. 40 p.
Sec. \(20 / 40 / 60 \mathrm{v}\). 2 amp. \(£ 2\) ea. P.P. 40 p .
L.T. TRANSFORMER (CONSTANT VOLTAGE). Prim. 200/240v. Sec. \({ }^{1}\)
\(100 \mathrm{~m} / \mathrm{a}\) E3. P.P. 50 p
\(100 \mathrm{~m} / \mathrm{a}\) £3. P.P. 50p.
L.T. TRANSFORMER. Prim. \(110 / 240 \mathrm{v}\). Sec. \(2 \times 32 \mathrm{v}\). (a 4 amp .
20v. (a 5 amp.: 15 v . (a \(1.5 \mathrm{amp} .: 7 \mathrm{~V}\). (a). 2.5 amp . E . P.P. 50 p .

20v. (a 5 amp .: 15 v . 1.5 amp .: 7v. (a) 2.5 amp . £3. P.P. 50 p .
L.T. TRANSFORMER. Prim. \(220 / 240 \mathrm{v}\). Sec. 13 v .
1.5 amp 65p. P.P. 15 p .
L.T. TRANSFORMER. Prim. \(115 / 240 \mathrm{v}\). Sec. 10.5 v .
at 1 amp. c.t \(28-0-28 \mathrm{v}\). at 2 amp . shrouded type. \(\mathbf{E 2}\).
P.P. 40 p

2500 watt. ISOLATION TRANSFORMER (CON-
STANT VOLTAGE). Prim. \(190-260 \mathrm{~V}\). 50 Hz . Sec. \(\mathbf{2 3 0 v}\). at 10.9 amps. \(\mathbf{£ 3 0}\). Carr. \(£ 2\)
H.D. STEP-DOWN TRANSFORMER. Prim. 200/240V
 \begin{tabular}{l} 
H.T. TRANSFORMERS. Prim. \(200 / 240 \mathrm{v}\). Sec. \\
\(300-0-300 \mathrm{v} . ~\) \\
\hline \(0 \mathrm{~m} . \mathrm{a} . ~\) \\
6.3 v . c.t. 2 amp £ .50 P.P. 40 p.
\end{tabular} \(300-0-300 \mathrm{v} .80 \mathrm{~m} . a .6 .3 \mathrm{v}\). c.t. \(2 \mathrm{amp} \mathrm{£1} 50\) P.P. 40
\(350-0-350 \mathrm{v} .60 \mathrm{~m} . \mathrm{a} .6 .3 \mathrm{v}\) c.t. 2 amp £1. P.P. 25 p. \(350-0-350 \mathrm{v}\). 60 m.a. 6.3 v . c.t. 2 amp . ER. P.P. \(22 / 240 \mathrm{v}\).
STEP-DOWN TRANSFORMERS: Prim. STEP-DOWN TRANSFORMERS: Prim. 22/240V.
Sec. 115 v . Double wound 500 w . f5. P.P. £1. 700 w . Sec. 115 v . Double wound 500 w . 5 . (metal cased with
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P.P. 50p 750W. £6. P.P.£1.
L.T. TRANSFORMER. Prim. \(110 / 240 \mathrm{v}\).
1 SA. (Shrouded type). \(£ 1.50\). P.P. 25p. HT/LT TRANSFORMER Prim. 240v. (tapped) Sec. 1.
\(500-0-500 \mathrm{v} .150 \mathrm{~m} / \mathrm{a} . ~ \mathrm{Sec} .2231 \mathrm{v}\). 5 amp. \(£ 2.75\) 500-0-500v. \(150 \mathrm{~m} / \mathrm{a}\). Sec. 2. 31v. 5 amp. \(\mathbf{~} 2.75\) P.P. 50p. Lyons) Input: \(190-260 \mathrm{v}\). Output \(240 \mathrm{v} . \pm 15 \% .124\)
K.V.A. \(\mathbf{f 6 0}\).

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Types 85p. P.P 5 p .
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UNISELECTORS (Brand new) 25-way
75 ohm. 8 bank \(\frac{1}{2}\) wipe £3-25. 10 bank
BLOWER FANS (S ail type) Type 1: Housing dia \(3 \frac{1}{}\) BLOWER FANS (Snail type) Type 1: Housing dia. \(3 \frac{1}{2} \mathrm{in}\).
Air outlet \(1 \frac{1}{4} \times 1 \mathrm{in}\) £2.25. P.P. 25p. Type 2 : Housing dia. 6 in . Air outlet \(2 \frac{1}{2} \times 2 \frac{\div}{2}\) in. £4. P.P. 50 p . Both types 115 / 240V. A.C. (brand new).
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SIEMENS/VARLEY PLUG-IN. Complete with transparent dust covers and bases. 2 pole c/o contacts 35p ea; 6 make contacts 40 p ea.; 4 pole c/o contacts 50p ea. 6-12-24-48v types in stock.
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core \(£ 24.20\) per 100 yds, 12 core \(£ 19 \cdot 80\) per 100 yds, 8 core \(£ 13 \cdot 20\)
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All poles rated at 250
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Items would include a malns operated, geared motor to drive ltems would include a malns operated, geared motor to drack
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\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|c|c|}
\hline OA2 & \(0 \cdot\) & qucs & & BL7(M) & 0 & & & & AL60 & 0.78 & L21 & 0.60 \\
\hline OR2 & 0.33 & 6BE6 & 0. & ¢L12 & 0.34 & 12AT7 & 0.20 & 0.95 & ARP3 & 0.35 & EC53 & 0.49 \\
\hline OZ4 & 0.44 & \(\therefore\) BG6G & 1.35 & \({ }_{6} \mathrm{~L} 18\) & 0-49 & 12AU6 & \(0 \cdot 38\) & 30 P 120.69 & ATP4 & 0.40 & EC54 & 0 \\
\hline 1 A 3 & 0.49 & \({ }^{3} \mathrm{~B}\) F & & Gila & \(2 \cdot 00\) & 12AL: & 0.21 & \(\begin{array}{lll}30 \mathrm{Pl} 16 & 0.31\end{array}\) & AZ1 & \(0 \cdot 40\) & EC86 & 0.59 \\
\hline \(145 g\) T & 0.48 & \({ }^{6} \mathrm{BJ} 56\) & 0.38 & 6LDI2 & 0.30 & & & 30P19/ & AZ31 & 0-48 & EC & 0.59 \\
\hline 1A7ti & 0.33 & ? \(\mathrm{BRK}^{\text {a }}\) & \(0 \cdot 60\) & 6LD20 & 0.55 & 12AV6 & 0.28
0.22 & \({ }_{30 \mathrm{P}}^{3} 10.65\) & \({ }^{\text {AZ }}\) +1 & 0.53 & EC92 & 0.34 \\
\hline \(1 \mathrm{B3GT}\) & 049 & \({ }^{\text {QRQ }}\) & 0.23 & 6N7(GT & 0.60 & 12AY7 & 0.80 & 30 PLL 10.57 & B:319 & 0.28 & ECC32 & 1.50 \\
\hline 1 C 2 & 0.50 & QRQT & 0.50 & 61P1 & 1.75 & 12 BAB & 0.30 & \(30 \mathrm{PLL1} 20.29\) & Blita & 0.50 & ECC33 & 1.50 \\
\hline \(1 \mathrm{L6}\) & 0.75 &  & 0.90 & 9P15 & 0.23 & \({ }_{1218 \mathrm{E} 6}\) & 0.30 & 30PLI3 0.75 & CL33 & 0.90 & ECC35 & 0.95 \\
\hline \(1 \mathrm{H5O}\) & 0.55 & \({ }^{\text {abra }}\) & 0.75 & 18P2x & 0.70 & \(12 \mathrm{BH}^{-}\) & & \(30 \mathrm{PLL1} 40.75\) & \(\mathrm{CVB}^{\text {c }}\) & 0.58 & ECC & 0.88 \\
\hline 1 L 4 & 0.14 & \({ }_{681887}\) & 1.40 & \({ }_{\text {fPLI2 }}\) & 0-28 & 12 SJGT & & \(30 \mathrm{PLL5} 587\) & CV63 & 0.53 & \(1 . \mathrm{CCR1}\) & 0.20 \\
\hline 1LD5 & 0.68 & \({ }^{\text {¢BW6 }}\) & 0.72 & 6479 & 0.44 & & & \(\begin{array}{ll}35 \mathrm{~A} 3 & 0.48\end{array}\) & CY988 & 0-10 & ccss & 0.21 \\
\hline ILNS & 0.86 & \({ }^{68 W}{ }^{\text {b }}\) & 0.50 & 687GT & 0.47 & 12K下\% & \[
\begin{array}{r}
0.35 \\
0.53
\end{array}
\] & \(\begin{array}{ll}3545 & 0.75\end{array}\) & CY1C & 0.55 & ECC83 & 0. 22 \\
\hline 1N5GT & 0.60 & \({ }_{68 \mathrm{BX}}^{1}\) & & & 0.55 & 12 K 7 GT & & 35050 & CY31 & 0.29 & ECC8: & 0.30 \\
\hline 1 RS & 0.28 & \({ }^{6 B 17}\) & 0.28 & \(\mathrm{b}_{6} \mathrm{~F} 7 \mathrm{~F}\) & \(0 \cdot 60\) & 1207 GTO & - 45 & 35 LGOT 42 & & 0.25 & ECCA & 0.34 \\
\hline 184
185 & 033 &  & 0.49
0.28 & \({ }^{1 s R} \mathbf{R}\) (M) & 0.75 & 12NATGT & T. 55 & \(\begin{array}{ll}35 \mathrm{~W} 4 & 0.23 \\ 3573 & 0.50\end{array}\) & DAC32 & 0.55 & ECC8t & 0.40
0.35 \\
\hline 185 & 0.22 & \({ }_{604}^{60} 4\) & 0.28 & esa & 0.35 & 12sc \({ }^{-}\) & & \({ }_{35 \mathrm{Z} 4 \mathrm{GT}}^{3584}\) & DAF & 0.36 & ECC88 & \(\begin{array}{r}0.35 \\ .48 \\ \hline\end{array}\) \\
\hline 104 & 0.44
0.80 & \[
\begin{aligned}
& 603 \\
& 1609
\end{aligned}
\] & 1.00 & \(6 \mathrm{6C7}\) & & 12897 & 0.38 & \({ }^{3524 G T}{ }^{3} \mathbf{2 4}\) & DC90 & & & 48 \\
\hline \[
\begin{aligned}
& \text { 1U5 } \\
& 2 \mathrm{D} 21
\end{aligned}
\] & 0.80
0.44 & \({ }_{6 C B 6}^{6 C 9}\) & 1.00 & dixa7( & 44 & 12815 & & \(3525 G T\)
50 BF
0.35 & & 18 & EC & \\
\hline 2 CKJ & 0.55 & 6 Cl 2 & 0.28 & \({ }_{68 \text { 6S7 }}^{68}\) & & \({ }_{12 \mathrm{NS}}^{12 \mathrm{~N}} \mathbf{}\) & 0.44
0.55 & \(\begin{array}{ll}50 \mathrm{C} 5 & 0.32\end{array}\) & DF96 & & ECF8 & 0.27 \\
\hline 3 A 4 & 0.36 & \({ }^{12}{ }^{\text {C17 }}\) & 1.00 & 6®K70T & & \({ }^{2 \times 80}\) & & \(50 \mathrm{CD6G}\) & \({ }_{1} 1863\) & & ECF8 & 25 \\
\hline 3B & 1.00 & - & 0.80 & BSQIGT & - 38 & 12ng: & & 2 & \({ }_{1} \mathrm{DHF}_{\text {His }}\) & 0.45 & ECF8 & 84 \\
\hline \(3 \mathrm{B6}\) & 0.19 & \({ }_{6}^{6 C}\) & 0.75 & 6U4GT & 0.70 & 1414 & 0.55 & \(0 \cdot\) & D117\% & 0.30 & & \\
\hline \(3 \mathrm{S4}\) & 0.49 & & 0.55 & 6U7G & - & 1487 & 0.75 & 50 & DH81 & 0.75 & & \\
\hline 3 396 & \(0 \cdot 55\) & \begin{tabular}{l}
6CLb \\
6CLsA
\end{tabular} & 0.46 & 6 V 4 & 0.19 & \[
\left\lvert\, \begin{aligned}
& 1+8 \\
& 18
\end{aligned}\right.
\] & 1.00 & & 0к32 & 0.33 & & 0.63 \\
\hline 354 & 0.28 & \({ }_{6 C M}\) & \({ }_{0} 0.75\) & & 0.17 & 19A05 & 0.42 & & L)K & 0.55 & & \\
\hline 5CG8 & 0.55 & \({ }_{\text {ficus }}\) & 0.75 & \({ }_{6}^{6 \times 69}\) & 0.27 & 19BGUG & &  & & & EC & \\
\hline 5 R 4 G & 0.70 & 6 CW 4 & 0.70 & 6 Y & & & 1.40 & boag 3. & & & ECH8: & 0.38 \\
\hline 5 T 4 & 0.30 & \({ }^{\text {fi }}\) & 0.60 & \({ }^{6 \times 6} 6\) & 0.65 & \({ }_{20 \mathrm{D}}^{19.9}\) & 2.00 & \(90 \mathrm{AV} \quad 3.38\) & DI & 0.38 & ECH84 & 34 \\
\hline 5 C 4 G & 0.30 & 60 CD 7 & 0.75 & 6y7c & 1.00 & 20 D 4 & & 1.78 & & \(0 \cdot 30\) & ECL80 & 0.28 \\
\hline 5 S 4 C & 0 & 6DT6 & 0.75 & TA \({ }^{\text {\% }}\) & 1.00 & \({ }_{20} \mathrm{H}^{2}\) & & 1 0 & DMT1 & 0.50 & ECL82 & \\
\hline 5 Y 3 G & 0.30 & & 0.75 & - 136 & 0.75 & 20L1 & 0.80 & \(90 \mathrm{Cl} \quad 0.59\) & DW \(+/ 350\) & & ECL. 83 & 0.52 \\
\hline 5Z: & - & 6EV & 0.75 & \(7 \mathrm{B7}\) & & \({ }_{20 \mathrm{P}}^{20 \mathrm{~L}}\) & 0.80
0.35 & & & 0.38 & ECL8 4 & 0.54 \\
\hline 5240 & 0.34 & 6 Fl & 0. & -58 & \(1 \cdot 00\) & \({ }_{20 \mathrm{P}}^{20}\) & 0.35 0 & \(150 \mathrm{C} 2 \quad 0.33\) & DY87/6 & & ECL85 & 0.54 \\
\hline 58497 & \(0 \cdot 38\) & \({ }_{6}^{\text {of6 }}\) & 0.35 & 7H7 & 0.55 & 20 & 0.80 & \(\begin{array}{ll}2158 G & 0.33\end{array}\) & DY802 & 0.30 & ECL86 & \\
\hline 6ABG & 0.55 &  & & iRT & 1.50 & & \({ }_{0}^{0.85}\) & \(\begin{array}{ll}301 & 1.00 \\ 302 & 0.83\end{array}\) & Esocc & 1.65 & & \\
\hline \({ }_{\text {cisar }}\) & 0.44
0.15 & \({ }_{6 F 14}^{6 F 12}\) & 0.40 & \({ }^{7} 9\) & 1.00 & \({ }_{25}{ }^{\text {afig }}\) & 0.38 & \(\begin{array}{ll}30.2 & 0.83 \\ 30.3 & 0.75 \\ \\ \end{array}\) & E80F & 1.20 & EF40 & \\
\hline 6AGS & 027 & \({ }^{6 F 15}\) & 0.65 & 724 & 0.80 & \({ }^{25 L 6 G}\) & \(0 \cdot 20\) & 3050.83 & E883 & 1-20 & EF42 & 0.33 \\
\hline batis & 0.50 & \({ }_{6 F 18}^{6818}\) & 0.55 & 9130 & 0.65 & 25 y & 0.38 & \(807 \quad 0.59\) & E92CC & & & \\
\hline 6A, \({ }_{\text {GAJ }}\) & 0.75
0.28 & 6F23
\(6 \mathrm{~F}_{24}\) & 0.6 & \(9 \mathrm{D7}\) & 0.40 & & & & El 180 F & & EFRO & \\
\hline 6AJS & 0.28
0.27 & \({ }_{6}^{6 \mathrm{~F} 24} 8\) & 0.51 & 10 C 2 & 0.65 & \({ }_{25 \% 5}^{252+4}\) & 0.33
0.60 & \(\begin{array}{ll}1821 & 0.5 \\ \text { อิ702 } & 0 .\end{array}\) & E182CC & 1.00 & EFr83 & 0 \\
\hline GAK6 & 0.60 & 8 F 26 & 0.28 & 10 Cl & 0.28 & 2826: & 0.70 & \({ }^{5763} \quad 0.50\) & E1148 & 53 & EF86 & 27 \\
\hline 6 AK 3 & \(0 \cdot 30\) & 6F28 & 0.60 & 1011 E & 0.55 & 2807 & 1.00 & 6060 0.30 & EAJí & & EF89 & 0.23 \\
\hline bALB & 0.12 & \({ }_{6 \mathrm{G} 6 \mathrm{~S}} 6\) & 0.30 & 10F1 & 0.50 & 30AJ & 0.65 & 71930.53 & & & EF91 & \(0 \cdot 17\) \\
\hline 6AM & 0.55 & 6G69 & \(0 \cdot\) & 10Fy & \(0 \cdot 65\) & 30 Cl & 0.28 & 74750 & Eabc8 & & EF92 & \(0 \cdot 30\) \\
\hline HANA & 0.48 & \({ }^{\text {baH8 }}\) & 0.75 & 10 Fl 18 & 0.55 & 300C15 & 0.58 & A1834 1.00 & EAC91 & & EF & 0.28 \\
\hline 6AQs & 0.22
0.94 & & & 10 LL 14 & 0.33 & 30 Cl 7 & 0.76 & A2134 & RaF42 & 0.48 & & \\
\hline GARS & 0.55 & 6H6GT & 18 & 10 LD & & \(30 \mathrm{Cl}{ }^{3}\) & 0.55 & \({ }^{\text {AC2PEN }}\) & EAF801 & & EFIB:3 & 0.25 \\
\hline 6 AR6 & 1.00 & 6JJGT & 0.29 & 10PL12 & & 30 Fs & 0.81 & 0.98 & FB3-1 & & EF184 & 0.27 \\
\hline 6 6as7 & 1.00 & \({ }_{6}^{6} 56\) & 0.20 & 10 Pl 3 & 0.54 & 30 FLL & 0.58
0.60 & AC2PENDD & FBC41 & & EFP94 & 1.20
0 \\
\hline gath
6 aUs & 0.30 & \({ }^{6 . J} 7(\mathrm{M})\) & \({ }_{0} 2.38\) & 10P14 & 2.00 & 30 FL 12 & 0.68 & & EBC81 & 0.29 & EH90 & 60 \\
\hline bavb & 0.33 & 6, JUa & 0.75 & P18 & 0.28 & \(30 \mathrm{FL13}\) & 0.50 & 0.38 & EBC90 & 0.30 & EK90 & 0.20 \\
\hline 6AW8 & 0.65 & 6K74 & 0.12 & 12A6 & 1.00 & 30 FL 14 & 0.66 & AC/PEN(7) & EBC91 & 0.28 & EL32 & 0.18 \\
\hline 6 AX 4 & 0.55 & 6 K 8 & \(0 \cdot 33\) & 12AC6 & 0.55 & :30L3 & 0.89 & 0.88 & EBF80 & 0.30 & ELI34 & 0.48 \\
\hline ab8G & 0.25 & \({ }^{6} \mathrm{Ll}\) & 2.00 & 12ADE & 0.80 & \(30 \mathrm{L15}\) & 0.55 & AC/THl 30 & EBF8 & \(0 \cdot 38\) & EL & \\
\hline 6 B & 0.18 & L5GT & 0.5 & 2 & \(0 \cdot 80\) & 301.17 & 0.65 & AC/TP 0.98 & EBF89 & & EL4 & \\
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\section*{mer IIOMI PE.AUDIOLC IUANTICHARI}

\section*{FREE-Audio IC Identichart}

Whether you are thinking of building an amplifier or you want some ideas on using i.c.s. in audio circuits, the Audio I.C. Identichart is exactly what you need. It gives comprehensive data on over 80 i.c.s. ranging fromlow level preamplifiers to hybrids rated at 50W. As well as data, the chart contains suggestions for using i.c.s. in mixers, tape recorders, record players, etc.

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\begin{tabular}{|c|}
\hline GEX541B1P2 ......f6 \\
\hline GEX541 D2P1 \\
\hline \multirow[t]{3}{*}{GEX541NB1P1F... E6.00 GEX541HP3F . . . . . 600} \\
\hline \\
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Edwards High Vacuum "Speedivac" model VSK1B range \(25 \cdot 760\) torr contact ratings 250 v .5 a . volume \(4.2 \mathrm{cu} . \mathrm{cm}\). max. working f6.20 Belling Delay hand reset L415 \(\quad £ 1 \cdot 10\) Stackpole min. rocker 125v.10a.250v. 5 5a
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6 watt (peak) Amplifier 240 v . AC, with inputs for Radio, Tape Recorder, freq fesponse \(80-12,500 \mathrm{~Hz}\), bass and treble controls, 2 speakers. Dimensions \(265 \times\) \(235 \times 580 \mathrm{~mm}\). Net weight 10 Kg . Idea for education seminars etc. \(\mathbf{£ 0 0} \mathbf{0 0}\) incl
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CAPACITORS


Daly Electrolytic 9000 uf 40 v .50 p ; Wego paper \(4 \mu \mathrm{f} 400 \mathrm{v} 60 \mathrm{p}\); Dubilie AC 35 p . TCCV13 Type 426100 uf 15 Jv . DC 50p; R.I.C. type 12971.8 uf 440 v MOTOR'S

GEC fractional \(1 / 12 \mathrm{hp} 230 / 250 \mathrm{v} 1 \mathrm{ph} 50 \mathrm{c} 2850 \mathrm{rpm} . . . . . . . . . . \mathbf{E S}^{2} 50\) carr. 67 p E.E. \(\frac{1}{2}\) hp 230v. 50 c 1 ph 50c. 1440 rpm compl ete with cap \(80 / 100 \mathrm{uf} 275 \mathrm{v}\).... \(\mathbf{5 1 3 - 0 0}\) 76813-393 Potter Instr. 110v. DC 4amp 0.2 hp . Cont. flange mounting precision tape transport motor (£80 value). \(\mathbf{£ 2 5} \mathbf{0 0}\) incl. car.
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Alrmax Type M1/Y3954 (3 blades) Cast Aluminium alloy impeller \& casing (corresponds to current type \(3965{ }^{1 \mathrm{ph}^{\prime}}{ }^{\text {"' }} 50 \mathrm{c} 230 \mathrm{v}\). 425 cfm free air weight \(9 \frac{1}{\mathrm{l}} \mathrm{bs}\) inct £21.00.
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The latest B.S.R. 8 Track cartridge Replay Deck. Ready to install in your Hi-Fi Stereo System.
This unit comes complete with Hi Gain Stereo Pre-Amplifier, 4-Programme Indicator Lamps. Track Selector Switch, al leads and plugs, etc. for 230 volt A.C. mains operation

\section*{5W \& 10W AMPS} 5Wonv£1.98 10W oniv \(£ 2.49\)
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Spacification:-
Nonimal Valts
Into 30hms
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Typical Distortion
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\$0\textrm{Hz}\mathrm{ to 30KHz 10Hz to 30KHz}
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The 5W matchbox sized amplifier will run satisfactorily from a 12 V car battery. Can also be used for portable voice reinforcement such as public functions where mains supply is not accessible. A small mains unit kit is avaiłable
Two amplifiers are ideal for Stereo. Complete connection details and treble, bass, volume and balance control circuit diagrams are supplied with each unit.
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A ready built unit, ready for connection to the I.F. stages of your existing FM Radio or Tuner. A tell tale light can be connected to show the presence of a Stereo transmission and correct operation.
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Supplied with all necessary instructions.

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5 W Amps
for
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\hline 07114472 & 10 & 4700 & 2.5 mps & 102 & 15 \\
\hline 07114682 & 10 & \({ }_{6300}^{680}\) & \({ }^{4} \mathrm{amps}\) & 102 & 17 p \\
\hline \({ }^{071} 071154723\) & 16 & \begin{tabular}{l}
3360 \\
\hline 400
\end{tabular} & \({ }_{3}^{2.4} \mathbf{4} \mathrm{mps}\) & 102 & 178 \\
\hline 07115682 & 16 & 68800 & \({ }_{5.8}^{3.8 \mathrm{amps}}\) & \(\underset{\substack{102 \\ 102}}{102}\) & \({ }_{220}^{17 p}\) \\
\hline 07115103 & 16 & 10000 & 7.9 amps & 2toz & \({ }_{27 \mathrm{p}}\) \\
\hline ( 071182828 & 63
10 & \(11000{ }^{2200}+11000\) & 5.8 amps & 302 & \({ }^{30}\) \\
\hline 07214173 & 10 & 160500 +16500 & \({ }^{10.6} \mathbf{1 3}\) amps & 302 & 37p \\
\hline 07215752 & 16 & \(7500+7500\) & 10.5 amps & \({ }_{302}\) & \({ }_{370}\) \\
\hline 07215113 & 16 & \(11000+11000\) & 13.8 amps & \(4{ }^{\text {+ }} 1\) & \({ }_{49}\) \\
\hline 007216502 & 25 & 500022005000 & \({ }^{2.2} 2 \mathrm{mmps}\) & 102 & \({ }^{15}\) \\
\hline 07216752 & 25 & 7500 +5000 & \({ }^{9.6}\) amps & 3102 & 370 \\
\hline 07217342 & 40 & \(3400+3400\) &  & 402 & P \\
\hline 07217502 & 40 & \(5000+5000\) & 12.0 amps & \({ }_{4}\) & 379 \\
\hline 0718681 & 63 & 680 & 2.4 amps & & \\
\hline 07218172 & 63 & \(1650+165\) & 7.8 amps & 302 & 37p \\
\hline \multicolumn{6}{|l|}{106 and 107 Series} \\
\hline 10614153 & & 15000 & 7 amps & 402 & \\
\hline 10615103 & \({ }^{16}\) & 10000 & 7 amps & +ox & 650 \\
\hline \({ }_{106} 171703\) & 25 & \({ }^{220000}\) & 17 amps & & E4.12 \\
\hline +10618193 & \({ }_{63}^{40}\) & 10000
15000 & \({ }^{12}\) amps & 7102 & +194p \\
\hline 10710222 & 100 & \({ }_{2200}\) & 10 amps & ¢ & 74p \\
\hline Type No. & Voltage & Capacilance & Weight & & Price \\
\hline 10215163 & 16 & \({ }^{16000}\) & \({ }^{800}\) & & 20p \\
\hline 10490003
102
16802 & 20 & \({ }^{39000}\) & \({ }^{1602}\) & & 30 p \\
\hline 10417562 & \({ }_{40}\) & 8000
5600 & 702 & & 25p \\
\hline 10490001 & 45 & 20000 & 1608 & & \({ }_{50 \mathrm{p}}\) \\
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\hline \multicolumn{2}{|l|}{Please calculate the weight of your order and include appropriate postage.} \\
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Parcels \\
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\hline & .... 67p \\
\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{2 C [뮤}} \\
\hline & \\
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\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{Ne,}} \\
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An aerosol spray providing a convenient means of produeing any number of copies of a Method: Spray copper laminate board with light sensitive spray. Cover with transparent film upon which circuit has been drawn. Expose to light. (No need to use ultra-violet.) Spray with developer, rinse and etch In normal manner. .. .. .. .. .. \(£ 1.00\) plus
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NEWER THAN NEW!!! Fibre Glass Board pre-treated with ont-senslitive lacquer enabing you to
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VCBO COLLECTOR TO BASE- 80 VOLTS,
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20 WATTS- 2 AMPS - 30 MHZ, FEATURES HIG
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Type SS15. These fine motors are easily reversed, starting and siopping in ess ar and holding torque of mechanical braking. Simple relay circult can be applied to give DC., to winding for a maximum holding torque o 60 Hz . 72 rDm . STEPPING. Holding torque at 60 steps per second- \(100 \mathrm{oz} / \mathrm{ln}\). Can be wired to glve 100 or 200 steps
 t" dis. Weight 6it ibs. BRAN FAN/ BLOWER Precision-built in Germany
Dynamically balanced main Dynamicatly balanced mains
unit (200/240) continuous unit ( \(200 / 240\) ) continuou
rated, reversible 60 MA rated, reversible 60MA o
run. Size: \(5 \hbar^{n}\)
dia. \(x\)
\(2 i\) deep. Back plate is tapped
for 4 fixing screws (supplled). Well under maker's price at \(£ 3\). \(P\). \& \(P\). 200 .
Similar unit to above but \(77^{\prime \prime}\) dia. \(\times 3^{\prime \prime}\) deep. \(\& 4 \cdot 50\). \(\frac{\text { P. \&P. 25D }}{8(P)}\) SMITHS RINGER-TIMER Reliable 15 minute times. spring wound (concurrent with time setting) \(15 \times 1 \mathrm{~min}\) divisions, approximately
divislons. Panel mounting with chrome divislons. Panel mounting with ehrom
bezel 3\(\}^{\prime \prime}\) dia. \(£ 130.150\). P . \(P\).


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"PEARED MOTORS \({ }^{\text {Gervalux" }}\) rpm geared motor. Type SD14,
\(230 / 250 \mathrm{v}\). \(A C\)., 22 ib/in. spindle. 1 st ciass condition
\(£ 7.50\) each. 50 p P. \& P. \(£ 7.50\) each. 50 p P. \(\&\) only as
Also limited number
above, BRAND NEW, \(£ 12.50\) Above, BRAND NEW, \(£ 12.50\) open rame shaded pole GEARED (Dural gear case)
240 AC., 28 rpm . NEW HIGH TORQUE, approx.
verall sIze: \(3{ }^{12} \times 3 t^{\prime \prime} \times\)

 110rpm with pressed steel gear case (similar to above
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\(5 \frac{1}{2} \mathrm{ram} .2 \mathrm{~m}^{2} \times 1 \mathrm{~m}^{2} \times 6 \mathbf{1}^{\prime \prime}\) high. \(240 \mathrm{v} . \mathrm{AC}\)



AMPEX 7.5v. DC MOTOR This is an uitra precision tape Motor designed for use in the recorder. Torque \(450 \mathrm{GM} / \mathrm{CM}\) retail load at 500 ma . Draws 50 ma on run. 600 mm . \(\pm\) speed
adjustment. internal AF/RF adjustment. Internal AF/R
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\(t^{*}\) dla spindie, motor \(3^{\prime \prime}\) dia. \(\times 11^{\prime \prime}\)
Original cost \(£ 16.50\). OUR P. 25 . Large §̧uantifies availab speclal quotations). Mu
75 p 日ach. FREEP. \& P.
"CROUZET" MOTORS Type 965
\(115 / 240 \mathrm{v} .50 \mathrm{~Hz} .48 \mathrm{w}\). Sloutly con-
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Type IMP Mk. 2. BRAND NEW and boxed. These well known timers are already in worid wide use and are
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SOLENOID This little unit gives vertical lift of approximately \(1^{\prime \prime}\) through
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by digits per unit. Robust and neat.
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PLUG-IN RELAYS by SCHRACK
(PERSPEX ENCLOSED)
OCTAL (2 c/0) 6 amp contacts at following voltages PRICE EACH 110 D.
2 A.C., 48 D.C., 48 A.C., 60 A.C. 60 Dmp contacts
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Base sockets for all sbove types 10 p .
Please add 10p towards P\&P on all ordurs.

From JAPAN. TAKAMISAWA Perspex enclosed rolays:
Type MO 308, 24V. DC. 600 ohms (4c/0), Complete with base Type MQ 308, 24V. DC.
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TANGENTIAL HEATER Slently drlven by shaded
pole Mycalex motor, powerfil and smooth running with aluminium impeller
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for 500 or \(1,000 \mathrm{w} . \mathbf{£ 1 . 8 0}\)
P. \&P. 30 p .
DOUGLAS TRANSFORMERS Full range in stock, SAE or phone for list please.


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 conductive properties. Heat res Istant, ideal for P.C. 's etc. THIS IS A
SPECIAL PURCHASE AND ONLY AVALABLE WHILE STOCKS LASTi SIzes: \(12^{2-} \times 12^{\prime \prime} ; 24^{4} \times 12^{\prime \prime} ; 24^{\circ} \times 24^{*} ;\) FULL SHEET \(43^{*} \times 37^{\prime \prime}\)
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'GOYEN' PRESSURE SWITCH Incorporating differential adjusiment between \(2^{\prime \prime}\) and
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OTEHALL Type
This swltch has a \(11^{\prime \prime} \times 15 / 3\)
fractlona! travel actuates. 6
PLEASE ADD 10\% FOR V.A.T. ON ALL
PRICES SHOWN INCLUDING P \& P
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S.F.M. \(-x 1, \times 10 . \times 100,0-75.750 \& 7500 \mathrm{kHz} \mathrm{dev}\) resp., P.M. -50 to 5000 pps at 1 to 30 usec
wIdth wldth int. or ext., 150 to 5000 Hz rep rate.
3 meters for; mad. and dev., frequency dlsComplete specification and price on

Open 9-12.30, 1.30-5.30 p.m. except Thursday 9-1 p.m.
noise. Pistance, Internal modulation at 1 kHz
impedance.
at up to \(90 \%\) depth, also external sine and at up to \(90 \%\) depth, also external sine and
pulse modulation. Built-In 5 MHz crystal
calibrator Separate R.F. and mod, meters.
P. \(\mathrm{F} 562 \mathrm{~A} / 3\) Oscliator and Detector Unit.

TM 577A for analysls and measurement of Radar Equipment. Frequency range 190 to 230 MHz
with crystai check polnts. Sweep width 0.5 to 5 MHz output pulae delay (a) \(85-175 \mathrm{HSec}\),
(b) \(0.7-1.4\) mSec with \(\times 1\) and \(\times 2\) multiplier and -2 , \(\times 1, \times 2\) multopller. Output
20 m
with \(\times 10\) multipller. \(£ 200\).

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TF 894 AUDIO \(T E S T E R\). Comblned A.F. Generator \((0-25 \mathrm{kHz})\), Output meter (up to
2 W . at 600,15 and \(3 \Omega\), and valve voltmeter (o-800V, with stepped and varlable attenua-
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takes the foltowing plugs-ins: \(X, Y\),
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pieligunurs
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134-P602t probe and current probe amplifier, \(1 \mathrm{~mA}-15 \mathrm{~A}\) p. \& p.i new and boxed \({ }^{\text {E12 }}\).
EQUIPMENT
162 wave form generator.
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M.O. for ET \({ }^{4336}\) TX (see description in AR88 SPARES. We hold the largent stock in U.K. Write for list. PHASE AUTO TRANSFORMER, wye
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\hline \multicolumn{2}{|l|}{TYPE MR．85P 4ilin．\(\times\) 4i in．fronts．} \\
\hline & \(\begin{array}{ll}10 \mathrm{~mA} & \cdots . . . \\ 50 \mathrm{~mA} & \cdots\end{array}\) \\
\hline & 100 mA …．．．\({ }^{33} 3.90\) \\
\hline & zooma
1 ampl \\
\hline &  \\
\hline &  \\
\hline &  \\
\hline &  \\
\hline  &  \\
\hline \({ }^{\text {S0，}}\) & 300 V A．C． C ． 83.90 \\
\hline \(10000-1004 \mathrm{~A}\) A 84.05 & 8 M eterimA 83.90 \\
\hline \({ }^{2004 \mathrm{~A}}\) … 5.54 .05 & VUMeter \\
\hline  &  \\
\hline  &  \\
\hline  & 20 mmp A．C．\({ }^{\text {a }}\) \\
\hline 5uA ……． 2399 &  \\
\hline \multicolumn{2}{|l|}{TYPE MR．52P \({ }^{\text {2fin．square fronts．}}\)} \\
\hline  & 10\％\％D．C．．．．． 52.50 \\
\hline  & 20v． \(\mathrm{B} . \mathrm{C}, \ldots\). ．\(£ 2.50\) \\
\hline \(100 \mu \mathrm{~A}\) ．．．．．．\(£ 3.00\) & 50V．D．C．．．．\(£ 2.50\) \\
\hline \(1001-0-1000 \mathrm{~A}\)（ 22.95 & 300V．D．C．．．£2．50 \\
\hline 万004A ．．．．．．\(£ 2.65\) & 15V．A．C．．．．．\(£ 2.60\) \\
\hline 1 mAA ．．．．．．．\(£ 2.50\) & 300V．A．C．．\(£ 2.60\) \\
\hline ธu1A ．．．．．．．．\(£ 2.50\) & \＄Meter 1ma ．\(£ 2.60\) \\
\hline 10 mA …．．\(£ 2.50\) & vo Meter ．．．\(£ 3.60\) \\
\hline  & 1 unp．A．c．＊\(£ 2.50\) \\
\hline 100 mA ．．．．．\(£ 2.50\) &  \\
\hline \({ }^{3} 00 \mathrm{maA}\) ． & 10 amp．A．C．＊\(£ 2.50\) \\
\hline 1 amp ．．．．．．．\(£ 2.50\) & 20 апр．A．C．＊ 22.50 \\
\hline 5 аиир．．．．．．．\(£ 2.50\) & 30 нrup．A．C．\(£ 2.50\) \\
\hline \multicolumn{2}{|l|}{TYPE MR．65P 3 3in．\(\times 3\) 3idin．frouts} \\
\hline 万̆uA ．．．．．． 83.70 & \\
\hline  & 20v．D．C．．．．．． 22.60 \\
\hline 100－1）－1004 A－ &  \\
\hline  & 300V：D．C．．．．\({ }^{\text {ex } 2.60}\) \\
\hline － &  \\
\hline  & 150V．A．C．．． 8280 \\
\hline  & Y000：A．C． 2.8880 \\
\hline  & 8 meter 1 mA 比 285 \\
\hline  & VU Meter ．．．\({ }^{23} 3.70\) \\
\hline  &  \\
\hline  & （1） \\
\hline \({ }_{15} 5\) amp．．．．．．\({ }^{2} 260\) &  \\
\hline  &  \\
\hline & \\
\hline 5v．11．c．\({ }^{\text {a }}\) ．．．． 82.60 & 30 ampl．A．C．\({ }^{\text {a }}\) \\
\hline \multicolumn{2}{|l|}{＂SEW＂EDUCATIONAL MET} \\
\hline & TYPE ED． 107 \\
\hline & Size overall \(100 \mathrm{~mm} \times\) \(90 \mathrm{~mm} \times 108 \mathrm{~mm}\) ． \\
\hline & A
quality
new
range of
moving \(\underset{\substack{\text { hight } \\ \text { coil }}}{\substack{\text { che }}}\) \\
\hline & ents idenl for \\
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\end{aligned}
\] \\
\hline \multicolumn{2}{|l|}{easily accessible to demonstrate internal working．} \\
\hline \multicolumn{2}{|l|}{\multirow[t]{6}{*}{}} \\
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\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|l|}{TYPE MR．45P} \\
\hline & \％2 65 & 10v．D．C．．．． & \\
\hline & & 20\％ & \\
\hline inma & 22．50 & 30．\({ }^{\text {a }}\)－ & \\
\hline & & & \\
\hline 30 JLA & \＆2．45 & 159．D．C． & \\
\hline 500－0－ & ¢2．40 & & \\
\hline & & ＊Meter 1 mad & \\
\hline & & U Meter & c2 \\
\hline 10 n & & 1 mmp ．A．C． & \\
\hline & 2． 2.40 & \({ }^{5}\) amp．A．C．： & \\
\hline 100 m A & & 10 amp ．A．C． & \\
\hline & & & \\
\hline & £2 & mp．A．C．\({ }^{\text {a }}\) & \\
\hline
\end{tabular}
＂SEW＂BAXELITE PANEL METERS
\begin{tabular}{|c|c|}
\hline &  \\
\hline & \(15 \mathrm{amp}, \ldots\).
30
30 \\
\hline &  \\
\hline &  \\
\hline &  \\
\hline & 50V．D．C．
150V．D．C． \\
\hline & \({ }^{1500 V}\) \\
\hline & 30v．A．C．＂． 22.65 \\
\hline  &  \\
\hline  &  \\
\hline \(1004 \wedge\) …．． 23.00 &  \\
\hline \(\begin{array}{ll}1000-100 \mu \mathrm{~A} & 83.00 \\ 82.70\end{array}\) & \\
\hline &  \\
\hline 500－10－8094 & 10 amp．A．C． 88.80 \\
\hline  &  \\
\hline 5 mA ．．．．．．．．\({ }^{\text {ce } 2 \cdot 60}\) & 30 amp．A．C． 22.60 \\
\hline 10 mA …．． 22.80 & 50 amp．A．C．＊ 22.80 \\
\hline  & VU Meter \(\ldots . .83 .65\) \\
\hline 100mA \(\ldots \ldots . .88 .80\) &  \\
\hline
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Two for \(\& 150 . P . \& P .75 p\).
All telephones complete with bell and dial. POTENTIOMETERS
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SUAYPĖS ALL BRAND NEW HIGH (1) \(3 \vee 9\) anp, \(6 \vee 9\) amp, \(12 \vee 9\) amp. Size \(3 \frac{1}{2} \times\)
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5 MHZ to 150 MHZ (Useful harmonics up to 1.5 GMZ ) up to 15 MHZ sweep width. Only 3 controls, preset RF level, sweep width and frequency. Ideal for \(10 \cdot 7\) or TV IF alignment, filters, receivers. Can be used with any general purpose scope. Full instructions supplied. Connect 6.3 V AC and use within minutes of receiving. All this for only \(£ 5 \mathbf{7 5}\). P. \& P. 25p. Suitable miniature transformer for 240 Volt operation \(£ 1 \cdot 25\).

Unless stated-please add \(£ 1 \cdot 50\) carriage to all units.

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\section*{Fantastic value in Test Equipment}


10 Channel EVENT RECORDER

Designed tor recording seauences of
up to ten difterent operations. e.g sequence of machine tool operation. switching sequences, etc. Record is
prespmted in the form of square "pulses When energised. pen moves by approxi-
mately 4 mm to the right of zero line mately 4 mm to the right of zero Che width 110 mm . Chart length 50 ft
nv. capaity 72 hours. Chart speeds
\(20.60-180-600-1800-5400 \mathrm{~mm} /\) hour Size \(160 \times 160 \times 255 \mathrm{~mm}\). Weight
Price complete with accessories

\section*{\(\mathbf{£ 5 2 . 0 0}\) \\ Supertester 680R.}

Buy it for what
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500 mp
511}

three channel HIGH SPEED RECORDER


Temperature Signal Producing
Pr
KHz and 500 KHz \(-200^{\circ} \mathrm{C}\) E11.95 sifnatist or terng. E5.95 Electronic Voltmeter

portable ac/dC RECORDING VOLTAMMETER Accuracy \(1.5 \%\) DC. \(25 \%\) AC. Measure ments ranges -AC and DC: 5-15-150 250.500 mA 250.500 V . DC only 150 mV . Frequenc 00 mm . Chart speeds \(20.60 \cdot 180.60\) 1800.5400 mm/hour Weigh:
£78.00

miniature pen recorder
Single channel high speed recorder Chart length 175 ft . Footage indicator Chart speeds iselected by push buttons der minute. Full deflection current 8 mA mpedance 800 ohms. Dimensions \(320 \times 340 \times 176 \mathrm{~mm}\) Ohms. Dimensions \(£ 55.00\)

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measurements of \(A C\) voltages and Specification
Measurement
10.25-100.250.500 \(\begin{gathered}\text { ranges:--Curen } \\ \text { Amps. }\end{gathered}\) Voltage 300.600 V Accuracy
lenger
60 mm
Overall
dimensions \(283 \times 94 \times 36 \mathrm{~mm}\). Weight 1.5 lb
\begin{tabular}{|c|c|}
\hline \(0^{2}\) a & \begin{tabular}{l}
wide fange \\
TRANSISTOR AUDIO \\
generator
\end{tabular} \\
\hline
\end{tabular} High stability low frequency generator. Basto
 means of a Schmitt trigel circuit. FREQUENCY
RANGES: 4 from 10 Hz to 100 kHz OUTPUT RANGES: 4 from 10 Hz to 100 kHz OUTPUT
VOLTAGE: 11 millivolt \(p-\mathrm{p}\) to 1 volt \(p-\mathrm{p} \pm 3 \%\) for voltace and square wave IMPEDANCE 1 K ohm DISTORTION FACTOR: \(0.2 \%\) for the sine wave
output for the lower frequency range \(<1.0 \%\) for outpur for frequency range. RISE TIME: \(<0.1\) the upper frequency range. RISE TIME <o. 0.
micraseconds tor the square wave and 0.3
microssconds in the upper frequency range microsegionds
WORKING miclosscands in the upper frequency range
WDRKRNG VOLTAGE. 9 voll. \(240 \mathrm{~mm} \times 140 \mathrm{mmH} \times x\)
 of the appoopriate accessones in can and phase seaumce. And there


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MARCONI TF867 STANDARD SIGNAL

GENERATOR
 \(10 \mathrm{c} / \mathrm{s}-100 \mathrm{Kc} / \mathrm{s}\). Complete with TM 6600 . Pulse adjustable between \(1.5 \mu \mathrm{sec}\). before and up to \(3,000 \mu \mathrm{sec}\)
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Range \(15 \mathrm{c} / \mathrm{s}-200 \mathrm{Kc} / \mathrm{s}\). Duration of square wave pulses between \(0.75 \mu \mathrm{sec}\) and \(40 \mathrm{~m} / \mathrm{sec}\). Square wave voltage AOV \(\overline{\text { AMPLITUDE MODULATOR TF } 102}\) PRICE £75.00
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\(100 \mathrm{Kc} / \mathrm{s}-300 \mathrm{MC} / \mathrm{s}\) Sine-wave from \(20 \mathrm{c} / \mathrm{s}-15 \mathrm{Kc} / \mathrm{s}\) and \(20 \mathrm{x} / \mathrm{s}-500 \mathrm{Mc} / \mathrm{s}\)
£35.00
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MARCONI TF2092 NOISE GENERATOR £295.00 MARCONI VHF SIGNAL GENERATOR TF 1145 450-1900 Mc/s \(£ 29500\)
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\(2 \mathrm{Kc} / \mathrm{s}\) - \(100 \mathrm{Mc} / \mathrm{s} £ 45 \cdot 00\)
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HEWLETT PACKARD 8690 SWEEP GENERATOR plus 8693 B Plug-in. \(3 \cdot 7-8 \cdot 3 \mathrm{GHz}\). £1,695.00.
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Dual Channel Transistorised DC-25 MHz at \(5 \mathrm{mV} / \mathrm{cm}\) 0.2 microsec. \(05 \quad 3 \% 5 \mathrm{X}\) Magnification extends sweep speed to 40 nanosec./cm. Sweep delay 180 nanosec.
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Rugged Transistorised fully portable Dual Channel DC-35 MHz at \(5 \mathrm{mV} / \mathrm{cm}\). As used by numerous government departments (c) COSSOR. The very latest Cossor 4000 Dual beam 55 MHz at \(50 \mathrm{mV} / \mathrm{cm}\) Trigger. SCOOP-ONE ONLY £425 DYNAMCO 7100 1Y2 \(71001 \times 2\) Oscilloscope. Dua channel with sweep delay, suitable for computer main-
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toys to 5 s delay. BRAND NEW \(£ 295\).


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K.G.M. Type 3015F 7 Segment display showing figures \(0-9\) plus decimal point. Character of 9 mm height. In 16 DIL case.

NEW LOW PRICE \(£ 1.40\) SN7447N BCD Decoder Driver \(£ 100\)

SINE COSINE POTENTIOMETER 47K Precision component by Pye. Model 2002 The assembly consists of three units mounted in one frane. Each unit conlains
two sine and two coside Dotentiometer sections. the sliders being ganqed together. Electrical connections. 2 end tans. Slider. Max, tortue: \(3 \frac{1}{2}\) Mechanical \(1 / \mathrm{P}\) : 30 r. P . DImensions: W in in. H. 5 in . D. \(7 \frac{1}{\text { in }}\). Wt. \(7 \frac{1}{1 / 2 \mathrm{lss} \text {. Ex equip- }}\) ment. Good condition. Price \(£ 5\). Carriage


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METER
A single beani instrument de efluent from a gas chromato graph, however the fast response
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CONTROLLER TYPE 990
Completely transistorised self-contained direct deflecting units for Suitable where a signal can be converted into d.c. Sensitivity 10 ohms per MV. Minimum F.S.D. 8 MV Cold function compensation. Calibrated
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\(\times 8 \frac{1}{2}\). weight 11 ibs. Mains suply \(100-260 \mathrm{~V}\). Control switching and thermo-couple connecti
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\section*{ASCOP DIGITAL ENCODERS \\ ype 504 A-8-001 Price £20. Type EDD8G Price \(£ 20\)}

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POWER SUPPLIES, IBM EX-COMPUTER HIGHLYSTABILISED, TRANSISTORISED LOW VOLTAGE POWER SUPPLIES.
\(\qquad\) NPUT and OUTPUT. Load regulation of \(1 \%\) or better. Low Avartable in the following iypes:


A BARGAIN ATLESS THAN HALF MANUFACTURERS PRICES.
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10 Volts 5 Amps
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20 Volts 4 Amps.
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6 Volts 11
11
18 EACH. P. AP, \(£ 2\)

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or a hand-driven generator and a direct reading ohmmeter

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Ideal for garages, this brand new instrument is used to display all ignition faults. Supplied complete with insiruction manuaa
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Ideal tor the measurement of low resistance. Low Current.
\(200 \mathrm{M} / \mathrm{A}\) at short circuit. Range \(1 \mathrm{~m} \Omega\) to 152 in 2 Ranges \(\pm 05\) milliohms or \(\pm 5 \%\) whichever is the greater. \(£ 20.00\).

SODECO IMPULSE PRINTING COUNTER 4Digl Decimal Counter 10c/second Electrical Reset \& Prin
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PHILIPS VALVE VOLTMETER
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Max. \(300 \mathrm{mV}, 1000 \mathrm{~Hz}-30 \mathrm{MHz}\)

\section*{to purchase some of the World's finest calibration instruments at savings of \\ PEN RECORDERS \\ BRAND NEW MINIATURISED STRIP CHART RECORDER BY RUSTRAK \\ of applied currents or voltages by a conlunuous distor
ion-tree \\ FREQUENCY CONVERTER MODEL B. 40}

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movement. scale callibrated 0 ol miltamp A.c. interna



\section*{SINGLE PEN \\ RECORDER}
by Record Electical. 3" chart, sensitivity
1 milliamp, chart speed 1 " \({ }^{\text {and }}\) 6" per hour

 packing and carriage.

LEEDS \& NORTHRUP STRIP CHART RECORDER
 Chart widh; 7 In . Chart speed 1 in. per hoirr. power supply: 120 50 So Hz

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TEN TURN 3600 ROTATION


50 KVA to 60 HZ power frequency converter. Fully overhauled Prime Mover: Electric Motor


HEWLETT PACKARD DIGITAL RECORDER MODEL 565A
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PRICE E85.00.
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In 4 ranges at 500 V Used Ior the measurement of components or
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ELECTRIC HAND VERIFIER
 02. ins. WIDERANGE OSCILLATOR TYPE 400C by DAWE FANS BY PLANNAIR
85V-3 Phase \(400 \mathrm{c} / \mathrm{s}-11.000\) rom. Type 1 PL41-234 PRICE 54.00 R.C. OSCILLATOR TYPE G432 by FURZEHILL SPECIAL OFFER SPECTRUM ANALYSER HEWLETT PACKARD \(8551 B\)
\(10 \mathrm{MHz}-12 \mathrm{GHz}\) and 851 B Extension to 40 GHz . With W/G Mixers and very little used Ex Calibration Lab. \(£ 3,950.00\).
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Dig
2 KV & \begin{tabular}{l}
DYNAMCO 2001 \\
Digital Voltmeter \(50 \mu \mathrm{~V}\) 2KV 0.05\% £175.00.
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\hline \multicolumn{2}{|l|}{DYNAMCO type 202} \\
\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{Long scale D.V.M. and Ratiometer. The 2022 is a high}} \\
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\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{potentiometric principle. It features a very high input}} \\
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\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{external scaling facility, seven operating modes and digital output.}} \\
\hline & \\
\hline \multicolumn{2}{|l|}{Scale . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 39999} \\
\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{Range
Resolution . . . . . . . . . . . . . . . . . . . . . . . . . 1 part in 40,000}} \\
\hline & \\
\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{Accuracy . . . . . . . . . Long term \(\begin{aligned} & 0.0025 \% \text { of F.S.D. } \\ & 0.01 \%\end{aligned}\)}} \\
\hline & \\
\hline & Optimum 0.0025\% F.S.D. \\
\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{Input Impedance \(\ldots . . . . . .0 .0025 \%\) of reading}} \\
\hline & \\
\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{}} \\
\hline & \\
\hline \multicolumn{2}{|l|}{MEGGER CIRCUIT TESTING OHMMETER} \\
\hline \multicolumn{2}{|l|}{For Measuring conductor resistance. By Evershed and} \\
\hline \multicolumn{2}{|l|}{Vignale. £22 50.} \\
\hline \multicolumn{2}{|l|}{\multirow[t]{3}{*}{\begin{tabular}{l}
BELL \& HOWELL \\
5-12 and 18 Channel U.V. Recorder \(£ 395.00\) 5-127 12 Channel \(£ 350\) 00.
\end{tabular}}} \\
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9 \& 10 CHAPEL ST., LONDON, N.W.I 01.7237851 01-2625125

\section*{STEP DOWN 240110v. AUTO TRANSFORMERS FOR} AMERICAN EQU,

\begin{tabular}{|c|}
\hline \multirow[t]{4}{*}{\begin{tabular}{l}
DAVENSET ISOLATION TRANSFORMERS \\
Pri. 10-0-200-220-240v. Sec. 240v. Centre tapped 1.2 kva , Conservatively rated. Size \(8 \frac{1}{4} \times 7 \times 8 \frac{1}{4}\) ins. Wot. 59 lbs . Open frame type, terminal connections. Fraction of maker's price. E17.00 carr. \(£ 1.00\).
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T.E.C. HEAVY DUTY ISOLATION TRANSFORMERS


RICH AND BUNDY. Pri. 220-230-240-250v. Sec. 265-270\(275 \mathrm{r}, 1400\) watts. Conservatively rated. Size \(8 \times 8 \times 7\) Ins
Terminal block connections. £17.00 carr. £ 1.00 .

ADVANCED COMPONENTS CONSTANT VOLTAGE Input \(190-260 \mathrm{v}\). Output \(230 \mathrm{v}, 150\) watis. Ty
 75 l .
Output 4 vu . 3 watis 75 pm carr. 25 p . Output 240 v . 30 watis en
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H.T. TRANSFORMERS BY FAMOUS

PARMEKO. Potted type. Prf. \(110-230-440 \mathrm{v}\). Sec. \(630-0-630 \mathrm{v}\) 105 mA . 5 v . \(4 \mathrm{a} ., 5 \mathrm{v}, 2 \mathrm{a} .53 .50\), carr. 50 p . Pri, \(110-220-240 \mathrm{v}\).
 WODEN. Pri. 230v, rame type table top connections, troplcallsed, \(\mathbf{£ 3 . 0 0}\) carr 50 p . Pri. \(220-240 \mathrm{v}\). Sec. 350 v .150 mA . 6.3 v .8 a .6 .3 v .3 a . ' C
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connections EA .00 carr . 75 p . Pri, \(220-240 \mathrm{v}\). Sec. \(350-0.350 \mathrm{y}\).


GRESHAM. Pri. \(220-240 \mathrm{v}\), Sec. \(710-0-710 \mathrm{v} .120 \mathrm{~mA}\). open rame type table top connections \(£ 2.75\) carr. 50 p . Prif. 110 50 p .
G.E.C. L.T. TRANSFORMERS
All Primaries \(220-240 \mathrm{v}\), Type 1 tapped. 63 - 68 -74v, 3 a . and
6.3v. 4a. terminal block connectlons. Unshrouded \(\mathrm{E} 3 \cdot 00\) S.3v. 4a. terminal block connections. Unshrouded \(£ 3.00\)
P.P. 50 p . Type 2 tapped. \(59-61-65-67-6 \mathrm{gv}\). 10 a . T blocks con nections. Unshruded \(\mathbf{E 6 0 0}\) carr. 75p. Type 3 tapped. 56

 iwice. Unshrouded, T block connections. £4.50 carr. 75 p Type \(610 \mathrm{v}, 2 \mathrm{a}\). and 50 v . 0.6a. T block connections. Un.
shrouded. \(\mathrm{E} \mid-50\) carr. 25 p . Type 715 v . 4 a . and 13 v . 6a. T
 2a. Twlee, unshrouded. £1.75. P.P. 30p.
PrI. 220-240y. S. \({ }^{2}\) ' CORE TRANSFORMERS Pri. 220-24.V. Sec.
6a. £2:25. P.P. 35 p


SPECIAL OFFER OF MULTI TAPPED L.T.
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TRANSFORMERS VERY CONSERVATIVELY RATED
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GRESHAM 'C.T. SMOOTHING CHOKES
GRESHAM 'C' core swinging types. \(7.5 \mathrm{~m} / \mathrm{h} .6 \mathrm{a}-75 \mathrm{~m} / \mathrm{n} 0.5 \mathrm{~s}\)
 G.E.C. \(15 \mathrm{~m} / \mathrm{ha}\). unshrouded fully troplcallsed \(£ 2.75 \mathrm{P} . \mathrm{P} .35 \mathrm{p}\) REDCLIFFE. Oll-filled types \(100 \mathrm{~m} / \mathrm{h}\). 2 at . \(£ 2.50 \mathrm{P} . \mathrm{P}\). \(45 \mathrm{p}, 130 \mathrm{~m} / \mathrm{h}\).
1.5 s . £1.50 P.P. 25 p . Mains filter chokes \(10 \mathrm{~m} / \mathrm{h} .2 \mathrm{a} .50 \mathrm{p} \mathrm{P} . \mathrm{P} .20 \mathrm{p}\) All above chokes 1 ohm res.
G.P.O. RELAYS 3000 TYPE \(100 \Omega 9\) 1.25 amp, Make contact.
G0D. P.P. \(10 \mathrm{D} .75 \Omega 3 \mathrm{M}, 1\) B. 1 CO Normal contacts. 40 D . B0p. P.P. \(10 \mathrm{D} .75 \Omega 3 \mathrm{M} .1 \mathrm{~B}, 1 \mathrm{CO}\). Normal contacts. 40 D
P.P. 10 p.
\begin{tabular}{|c|c|}
\hline \multicolumn{2}{|r|}{H.T. TRANSFORMERS} \\
\hline \multicolumn{2}{|l|}{\multirow[t]{4}{*}{PARMEKO. PrI. 240v. Sec. \(250-0-250 \mathrm{v}\). \(50 \mathrm{~m} / \mathrm{a} .6-3 \mathrm{v} .1 \mathrm{a}\), E1-25. P.P. 35p. Size \(4 \times 3 \times 2 \ddagger\) Ins. GARDNERS. 'C' core. PrI. 240v. Sec. \(300-0-300 \mathrm{v} .66 \mathrm{~m} / \mathrm{a}\). 6.3 v . 4a. E1-50. P.P. 35p. Size}} \\
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\hline \multicolumn{2}{|l|}{\multirow[t]{3}{*}{}} \\
\hline & \\
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\end{tabular}
T.C.C.-DUBILIER OIL-FILLED BLOCK CAPACITORS





TUBULAR MOTOR START CAPACITORS



 AMERICAN OIL-FILLED BLOCK CAPACITORS


 \(\frac{\text { wkg. 35p. P.P. 15p. }}{\text { H. }}\)
\[
\begin{aligned}
& \text { H.T. SMOOTHING CHOKES } \\
& \text { types. } 5 \mathrm{~h} .500 \mathrm{~m} / \mathrm{a} \text {. } £ 3 \cdot 00 \mathrm{carr} .50
\end{aligned}
\]

Parmeko potted H.T. SMOOTHING CHOKES
\(£ 2.00\), 5 h. \(500 \mathrm{~m} / \mathrm{a}\). \(£ 3.00\) carr, \(50 \mathrm{p}, 10 \mathrm{~h} .300 \mathrm{~m} / \mathrm{a}\) \(£ 2.00\) carr. 30 p. \(10 \mathrm{~h} .180 \mathrm{~m} / \mathrm{a}\). \(£ 1.50\) earr. 3 ( \(\mathrm{p} .15 \mathrm{~h} .180 \mathrm{~m} / \mathrm{a}\). \(£ 2.00\)
 earr. 20p.
A.C. GEARED MOTORS BY FAMOUS MAKERS 230/250v. 50 cyeles Inductlon type. \(4 / 2\) r.p.m. Cont. rating 5 lb ins.
Rlght angle worm drive. Overall size 7 ins. Dia. 3 ins. Spindle ength 3 ins. Dla. \(\frac{1}{4}\) in. \(\mathbf{E 4} \cdot 75\) carr, 45 p . Gear motors 50 V . D.C.
Shunt wound. Cont. rating. \(34 \mathrm{r} . \mathrm{p}, \mathrm{m}, 2 \mathrm{ib}\). Ins. Right angle worm


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HP. 1/35 A.C. 115v, 50 Cycles. RPM 137. Torque 9 In ibs.
Ratio 10-1. Puiley Drive. Complete with Control Box contal Ratlo 10-1, Puiley Drlve. Complete with Control Box contain
ing Capacltor. Onlof Swhith. Mlero switch reversing con ne=tions. Ideal for electric door systems. \(£ 10.00\) carr. \(£ 1\).

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This is our Bailey/Burrows Stereo pre-amp front end. We think it is the best engineered kit of the best pre-amp circuit available, and there is a back end/tone control unit of similar advanced design to go with it which is only \(1 \frac{1}{2}\) " deep so it fits almost anywhere, but of course it's at its best in a Hart universal amplifier metalwork with a couple of Hart Bailey 30 watt power amps to keep it company. That's a recipe for real \(\mathrm{Hi}-\mathrm{Fi}\) with electronics you'll be too proud to cover up.
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SAFETY ISOLATING TRANSFORMERS Pilmary \(120 / 240\) Volts. Secondary \(120 / 240\) Volts. Centre tapped with

Interwinding Screen.
(WATTS) WEIGHT.
\begin{tabular}{|c|c|}
\hline (WATTS) & LB. \\
\hline \({ }^{60}\) & 产 \\
\hline 100
200 & \({ }_{8}{ }^{\text {\% }}\) \\
\hline 250 & 13! \\
\hline 350 & 15 \\
\hline \(\begin{array}{r}500 \\ \\ \hline 000\end{array}\) & \({ }_{38}^{198}\) \\
\hline 2000 & 60 \\
\hline 3000 & \({ }^{85}\) \\
\hline 000 & 173 \\
\hline
\end{tabular}

SIZE CM
\(\qquad\)

TYPE
NO.
149
150
151
152
153
154
156
158
159
160
10
The above are aliso
On application.

MINIATURE \& EQUIPMENT TRANSFORMERS Pri. 240 Volts with interwinding Screen.
VOLTS MA WT. SIZECM.
\begin{tabular}{|c|c|c|c|c|}
\hline VOLTS & mA & & SIZE CM, & TYP \\
\hline 3-0-3 & 200 & Lb. & \(2.8 \times 2.6 \times 2.0\) & 238 \\
\hline 0-8, 0-6 & 500, 500 & & \(4.8 \times 2.9 \times 3.5\) & 234 \\
\hline 0-6, 0-6 & 1000, 1000 & 12 & \(6.1 \times 5.8 \times 4.8\) & 212 \\
\hline 9-0-9 & 100 & & \(3.9 \times 2.6 \times 2.9\) & 13 \\
\hline 0-9, 0-9 & 330,330 & I & \(4.8 \times 2.9 \times 3.5\) & 235 \\
\hline 0-8-9, 0-8-9 & 500, 500 & 1 & \(6.1 \times 5.4 \times 4.8\) & 207 \\
\hline 0-8-9, 0-8-9 & 1000, 1000 & 13 & \(7.0 \times 6.4 \times 8.1\) & 208 \\
\hline 15-0-15 & 40 & 1 & \(2.8 \times 2.6 \times 2.0\) & 240 \\
\hline 0-15, 0-15 & 200, 200 & \(\frac{1}{4}\) & \(4.8 \times 2.9 \times 3.5\) & 236 \\
\hline 20-0-20 & 30 & + & \(2.8 \times 2.6 \times 2.0\) & 241 \\
\hline 0-20, 0-20 & 150, 150 & 1 & \(4.8 \times 2.9 \times 3.5\) & 237 \\
\hline 0-15-20, 0-15-20 & 500, 500 & 2 & \(7.0 \times 6.7 \times 6.1\) & 205 \\
\hline 0-20, 0-20 & 300, 300 & 1\% & \(5.1 \times 5.8 \times 4.8\) & 214 \\
\hline 20-12-0-12-20 & 700 (DC) & 1 & \(7.0 \times 6.1 \times 5.1\) & 221 \\
\hline 0-15-20, 0-15-20 & 1000, 1000 & 2 & \(8.3 \times 7.7 \times 7.0\) & 206 \\
\hline 0-15-27, 0-15-27 & 500, 500 & 2 & \(8.3 \times 7.0 \times 7.0\) & 203 \\
\hline 0-15-27, 0-15-27 & 1000, 1000 & \(3 \frac{1}{7}\) & \(8.9 \times 7.7 \times 7.7\) & 204 \\
\hline & AUT & & NSF & \\
\hline
\end{tabular}

AUTO TRANSFORMERS



20/240 VOLTS TO 115 VOLTS WITH LEAD AND 115 VOLT SOCKET VA WOIGHT VOLTS WITH LEAD AND 115 VOLT SOCKET


LOW VOLTAGE TRANSFORMERS PRIMARY 200/250 Volts. SECONDARY 12 and 12 Volts.
AMPS
WEIGHT
SIZE CM.


30 VOLT TRANSFORMERS
\begin{tabular}{|c|c|c|c|c|}
\hline \multicolumn{5}{|l|}{\multirow[t]{2}{*}{30 VOLT TRANSFORMERS
PRIMARY 200/240, SECONDARY 12, \(15,20,24,30\)}} \\
\hline & & & & \\
\hline AMPS & WEIGHT & SIZE CM. & TYPE & PRICE \\
\hline & & & & \\
\hline \(\frac{1}{1}\) & \({ }^{1 \frac{1}{4}}\) & \(6.1 \times 5.8 \times 4.8\) & 112 & 1.20 \\
\hline 1 & 2* & \(7.0 \times 6.7 \times 8.1\) & 79 & 1.64 \\
\hline 1 & \(3 \pm\) & \(8.9 \times 7.7 \times 7.7\) & 3 & \(2 \cdot 45\) \\
\hline 3 & 4 & \(9.9 \times 8.3 \times 8.6\) & 20 & 3.00 \\
\hline 4 & 8 & \(9.9 \times 9.6 \times 8.6\) & 21 & 3/55 \\
\hline 5 & \(6 \frac{1}{2}\) & \(12.1 \times 8.6 \times 10.2\) & 51 & 4.40 \\
\hline 8 & & \(12.9 \times 9.3 \times 10.2\) & 117 & 5.28 \\
\hline 8 & 12 & \(12.9 \times 11.8 \times 10.2\) & 88 & 8.80 \\
\hline 10 & 13: & \(14.0 \times 10.2 \times 11.8\) & 88 & 8.36 \\
\hline
\end{tabular}


50 VOLT TRANSFORMERS
PRIMARY 200/240, SECONDARY 19, 25, 33, 40, 50
AMPS WEIGHT
\begin{tabular}{cc} 
SO & \\
TYPE & PRICE \\
No. & \(\neq\) \\
100 & 1.60 \\
103 & 1.35 \\
104 & 3.25 \\
105 & 4.40 \\
108 & 5.48 \\
107 & 8.65 \\
118 & 11.27 \\
119 & 14.15
\end{tabular}

BRIDGE RECTIFIERS
 25 p
25 p
28 p
30 p AMPS WEIGHT SIZE CM. \({ }^{2}\) TYPE PRICE POST 600
\(\qquad\) FOUR AMP
\begin{tabular}{|c|c|c|}
\hline & Volts & \\
\hline 200 & ', & \\
\hline 400 & " & \\
\hline 600 & , & \\
\hline 800 & & \\
\hline
\end{tabular} SIX AMP 65 p
78 p
80 p
99 p
c 1.00
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219.75
\(0 V^{\prime}\) into 600 S .

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For T. T. Line Defection Cirenit
Yebo Timiv. D.C. or 1500 peak
\&1. 80
MINIATURE WIRE ENDED SILICON
RECTIFIERS
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline 1 N 4002 & 100 p.i.v. & 1.4 & -. & . & . & & 07 \\
\hline 1 N 4004 & 400 p.i.v. & \({ }_{1}\) & . & \(\cdots\) & & & 08 \\
\hline \(1 \times 4006\) & 800 pi.is. & 1A & . & & \(\cdots\) & & 0.12 \\
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4 pole， 3 way－2 pole， 4 way－\({ }^{2}\) pole， 4 way－ 4 pole， 3 way－ 2 pole， 4 way－ 3 pole， 4 way－

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 ST 22 p DP
swe tipe． 13 amp self－fixing into an oblong hole．
size approximately \(1 \mathrm{ln} . \times\) tin．， 9 p each SLIDE SWITCHES

B．Ditto as above but for printed circuit 7 y each． sub Ministure 8Hide 8witoh．DPDT 19 mm （4＊
approx．）between fixing centres． 14 p each or 10 tor \(21 \cdot 26\).
DOUBLE LEAF CONTACT
 Very ellght pressure closes both
contacts．8p each， 10 for 72 p
Plastle push－rod suitable for operating．
6p esch． 54 p for 10 TELEPHONES Complete as illustrated．Save your
legs，time and temper．simply by lera，time and temper．simply by
putting in some telephones．Ex．
G．P．O．not new－but guaranteed in cood condition and serviceable． Supplied with diagram and instruc－
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I REV．PER MINUTE MOTOR WITH GEAR－BOX
Made by the famous Chamberlain \＆Hookham led．These


MINIAT URE SEALED RELAY
 high and it＇s a double change over，we this at \(3 / 5\) amps．The coll resistance is 600 ohms and 9.12 volt will close it．Ideal for modele and miniaturised equipment． with base．Price 28 p including base． METAL CHASSIS

former and／or valve holdera also some in \(^{n}\) holes for controls，
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Units made by Delco． 6 bladed \(5^{\prime \prime}\) dia．fan inside heavy
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 but can be run from A．C．up to 30v．The higher the voltage


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tronically changes Electronlcally changes speed
from approximately 10 reve from approximately 10 revs．
to maximum．Full power at ali
speeds by finger－tip control speeds by finger－tip control．
Kit includes all parts，case，
everything and
 BAKELITE INSTRUMENT
size approx． \(6 \dot{t}^{* *} \times 3 \xi^{* *} \times 2^{n}\) deep With
brass inserta in four comers and bakelite brass inserts in lour cormers and bakelite
panel．This Is a very atrong case suitable
to house fastruments and special riga etc． to house instrumentr and special rigs，etc ISA ELECTRICAL PROGRAMMER


Clock by famous maker with 15 amp．on／off switch．Switch Independent 60 minute memory jogger．A beautiful unit Price \(82 \cdot 15+\)
bezel \(83 p\) extra．


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\section*{HIGH ACCURACY}
as probe．Deaigner claims tempersture contro to within \(1 / 7 \mathrm{th}\) of a alims tempersture ocontrol
power pack \(46 \cdot 15\) ． TREASURE TRACER
Complete Kit（except wooden battens）to
makee the metal detector as the circuit in make the metal detector as the circuit in
Practical Wireless August issue． \(\mathbf{8 3 . 8 5}\)
pluas 20 p post and
AUTO TRANSFORMER Primary \(220-240 \mathrm{v}\) ．Secondary \(110-\)
120 v ．Well built and varnish impreg． ated． 250 watt intermittant rating． Size approx． \(34 \times 3 \times 3\) in．\(\times 3\) ．
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 ELECTRIC TIME SWITCH
 Made by smilthe these are A．C．mains operated．NOT e built into box with 13A socket． 2 completely adjurtable ime periods per 24 hours， 5 amp changeover contacts whl switch circuit on or off during these periods． 82.75 post and ins．，23p．Additional time contacta 55 p palr．



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This is a drum type timing device，the drum being
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woition． They are also arranged to allow－ 2 operations per switch
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 TANGENTIAL HEATER UNIT Thls heater unit is the very lateat type，most efficient．and quet running．Is as fited in more．We have a few only Comprises motor，
impeller； 2 kW ，ajement anvil kW ．element allowing impeller； 2 kW ，ajement gnvilikW．element allowing
switchlng 1,2 and 3 kW ．and with thermal safety


Don＇t mlse this．Control switch 44D．plus VAT P．\＆P． \(40 \%\)

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This unit ig self contaimed and on wheels．It stands approx． 6 ft high and 3 ft aquare．
On the front panel is a Variac，Voltmeter，a 60 second Timer as well as the normai On the front panel is a Variac，Voltmeter，a 60 second Timer as well as the norinal
overlood trip on off switch and cut outs etc．The transormer itself if oll flled and overload trip on／off switch and cut outs，etc．Fimary so all voltages up to 35 KV are
rated at 7 KVA 30 KV ．The varlac is in the primarity avallable．We beiieve the normal use for such a unit would
flash tester． 1 only－not new but in good order．Price sige．


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disc lighting effect，etc．，etc． 82.05 plus 20 p post and ineurance

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Just hat you need for work bench or lab． \(4 \times 13 \mathrm{amp}\)
soekets in metal box to take gtandard 13 amp fused
pluge and on／off switch with neon warning light Bup hred up ready to work，\(£ 2 \cdot 48\) plue 26 p \(\mathbf{P}\) ．\＆ module form，each ready built complete
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CAR PANEL SWITCH Our Ref．No．SO1．Arco made．
Has long fat ended toggle black
and chrome fnle and chrome finish．Rated 2 A．at 250 v ．and is double pole on／an．
Listed at 45 p．Our price \(2 R \mathrm{p}\)

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Ref．No．SO3．Again a fiat ended toggic．Made oy Arrow． auto arerials．reversing motors etc． 30 p each．

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For driving a bilge pump and similar applicatlons．This motor we understand develops on \(t\) H．P．It is extremely
powerfil and although rated at 6 v ，this operates up to powerful and although rated at 6 v ，this operates up
12 ve for short periods with very much increased power， （probably at least \(\$\) H．P．）We understand that from the
makers they cost over 5. At 82.20 each phus 25 p post on makers they cost over 25 ．At \(22 \cdot 20\) each plus \(25 p\) post on

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These are variable voltage tranaformers．Britloh made by
the famous Zenith Co．Fully enclosed for bench use and No 100 LM．220－240v．A．C．output 0.240 v ． Thls model is listed at over 220 ．We have a limited quantity only，absolutely brand new，still in maker＂s cartons，offered
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Made for Admiralty． 24 volt D．C．input， 240 v ． 50 cps． output．Admiralty rating 80 watta but we have tested this
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TRIGGER MODULE MY 5004
This produces pulaes for phase
control triggering，it has two isolated outputs，so one thyris－
tor or two thyristors in tor or two thyristors（in
separate arms of bridge）may be controlled by one module．
The timing circult is aynchro－
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9V GRAMOPHONE UNIT
Battery operated on unit plate 2 speed auto－ato
BUY TIME SLOT METER
Made by Sangamo Weston．3 types，one for each coln
2 pt ， 5 p or 10 p ．Price 81.75 each plus 25 p port and ins． 4 STATION TRANSISTORISED

\section*{INTERCOM}

Solld state three transistor printed cct．mater and tree sub station push button／press talk system． 200 mW output， power pack approx \(3 \times 14 \times 4\) in Price 88.50 plus 20 p ． PHOTO ELECTRIC KIT
Contains photo cell，relay，transistor and all parts to make AC／DC MILLIAMMETERS 3 RANGE Moving Iron mirror scale laboratory instalate．Ranges
 GALVOMETER 7－0—7 UA F．S．D．
Moving coil precision laboratory Inatrument of extremely \(6 t \times 24 \times 2\) in．Price 87.50 ．

\section*{ACOS．＇\(G\)＇METERS}

For use with tranaducers and accelerometers．These are precision instruments they measure＂ g ＂in three ateps
\(0-10,0-100\) and \(0-1000\) directly on a large clear meter scale 0－1．Two models available：－Standard model（IDOOL inbuilt clrcuit with relay to trip the external circuit（trip level is adjustable by a control which is virtually linear
with the meter scale）．The trip load may be up to 2 a．
Once the circult has been tripped it can be restored by a YOLTAGE CHATton．Price of this model is \＆18 YOLTAGE CHANGING TRANSFORMERS
MADE BY PARMEKO
Upright mounting，fully shrouded and with terminal blocks
for input and output．For changing mains voltage，ideal for input and output．For changing mains voitage，meal
for working low voltage equpment from \(230 / 240\) malna
and for increasing voltage due to losese fal long leads．Voltage and for increasing voltage due to loses in long leads．Voltage
up or down between \(190-250 \mathrm{v}\) ． 250 watts．Prlce \(\mathbf{~} 1.65\)

\section*{DOOR OPENING OR PLATFORM}

\section*{ROTATING MOTORS}

Very powerful motors eatimated rating at if H．P．Reversible with gearbox and＂V＂belt drive wheel of \(7^{\prime \prime}\) diameter．
These are by a French maker using trade name LUXOR and name plated Moteur Asynchrome．Capacitor start but
with 6 connections．At the time of writing this we haven＇t connection diagram the if any reader has mist thls motor before and can tell us how to connect and anything about it we would he obliged．Price of motor which welghs approx．
15 lbs i \(\& 10\) ． SPRING RETURN WAFER SWITCH
As used in intercom and other similar equipment a two wafer 6 pole 3 way awtch，opring return from centre position
when turned clockwise and permanent off or on when turned anti－clockwise．Prite 55p each．
A．C．B UZZERS
2v．Fix theae into a box which will resonate and they will
ive a loud piercing note．Suitable for alarms or aignal． PUSH ON TAG CONNECTORS
These are being increacingly used on cara and domestic betag．We offer these at attractive prices． 10 for 10 p ．
50 for \(40 \mathrm{p} \ldots 100\) for 70 p and 1,000 for \(\$ 5\) ．
WALL THERMOSTATS

This Month＇s Snip．Made by the famous Smiths
Instrument Co．called Colourstat．Wall mounting
Adjuatable by slider（lockable）and may be set to control temperatures from around ireezing through （frost）（warm）（very warm）etc．The thermostat． mains voltage and is ideal for liviny room，bed－
room and greenhouse etc．Price \(\boldsymbol{£ 1} \cdot \mathbf{6 5}\) ．Don＇t misa

Where postage is not stated then orders
over \(£ 5\) are post free．Below \(£ 5\) add 30 p． S．A．E．with enquiries please．

J．BULL（ELECTRICAL）LTD．
（Dept．W．W．）7，Park Street，Croydon，CRO 1YD Callors to 10213，Tamworth Road．Croydon

\section*{(P) 1.L_P. (Electronics) Ltd}

\section*{100 WATTS!}

\(\star\) NO EXTERNAL COMPONENTS
\(\star\) MECHANICALLY \& ELECTRICALLY ROBUST \(\star\) INTEGRAL HEATSINK
\(\star\) HERMETICALLY SEALED UNIT
\(\star\) ATTRACTIVE APPEARANCE
\(\star\) LOWCOST
\(\star\) BRITISH BUILT
\(\star 100 \times 105 \times 25 \mathrm{~mm}\)

With the development of the HY200, ILP bring you the first COMPLETE Hybrid Power Amplifier.
COMPLETE: because the HY200 uses no external components!
COMPLETE: because the HY200 is its own heatsink!
By the use of integrated circuit technique, using 27 transistors, the HY200 achieves total component integration. The use of specially developed high thermally conductive alloy and encapsulant is responsible for its compact size and robust nature.

The module is protected by the generous design of the output circuit, incorporating 25 amp transistors. A fuse in the speaker line completes protection.

Only 5 connections are provided, input, output, power lines and earth.
Output Power: 100 watts RMS; 200 watts peak music power into \(8 \Omega\)
Input Impedance: \(10 \mathrm{~K} \Omega\)
Input Sensitivity: ODb ( 0.775 volt RMS)
Load Impedance : 4-16 \(\Omega\)
Total Harmonic Distortion: less than \(0.1 \%\) at 100 watts typically \(0.05 \%\).
Signal: Noise: Better than 75 Db relative to 100 watts
Frequency response: \(10 \mathrm{~Hz}-50 \mathrm{KHz} \pm 1 \mathrm{Db}\)
Supply Voltage: \(\pm 45\) volts
APPLICATIONS: P.A., Disco, Groups, Hi-Fi, Industrial.
PRICE: \(\mathbf{£ 1 4 . 9 0}\) inc. VAT \& P \& \(\mathbf{P}\)
Trade applications welcomed

\section*{DP A-P.(Electronics) Ltd}

\section*{SECOND GENERATION 25 WATT HYBRID}


\section*{NEW HY5 PRE-AMPLIFIER}

Unchallenged for two years, the HY5, our unique multifunction preamplifier/tone hybrid, has been brought into line with the advancements in our power hybrids.
Like the HY50, the new HY5 has no external components \& has been redesigned to run off a split power line with improvements in signal/noise, overload capability \& reduced distortion. The output has been increased to match the power module (Odb). and to share the same power supply.
Overall size is reduced by the use of a new thin film circuitry while the device still retains all the functions of the eariler device.
When combined with the HY50 \& power supply only potentiometers are required to complete a simple mono amplifier with input \& output facilities expected to be found on Hi-Fi amplifiers.
The combination of two HY5's two HY5O's sharing a common power supply (PSU50) are linked by a balance control to form a complete stereo system.
INPUTS
Magnetic Pick-up 3 mV (within Idb RtAA curve)
Ceramic Pick-up up to 3 mV .
Microphone 10 mV .
Tuner 250 mV .
Auxiliary 3.100 mV .
Input impedance \(47 \mathrm{k} \Omega 1 \mathrm{kHz}\)
OUTPUTS
Tape 100 mV .
Main output. Odb \((0.775\) volts \()\)
ACTIVE TONE CONTROLS
Treble \(\pm 12 d \mathrm{~b}\) at 10 kHz
Bass \(\pm 12 \mathrm{db}\) at 100 Hz
OVERLOAD CAPABILITY (equalization stage) 40db on most sensitive inpu
OUTPUT NOISE LEVEL (below 10 mV magnetic input) 68 db .
DISTORTION \(0.05 \%\) at 1 kHz .
SUPPLY VOLTAGE \(\pm 16-25\) volts
SUPPLY CURRENT 15 mA .
Price \(£ 4.51\) mono \(£ 9.02\) stereo
Price inclusive of VAT \& P \& P.

\section*{POWER SUPPLY PSU50}

The new PSU50 has a low profile look being only \(2 \frac{1}{4}\) inches high and can be used for either mono or stereo systems. SPEC.
OUTPUTVOLTAGE \(\pm 25\) volts.
INPUTVOLTAGE 210-240volts.
SIZEL. 70 D. 90 H .60 mm .
Price £5.23.
Price inclusive of VAT \& P \& P.

\section*{APPOINTMENTS VACANT}

DISPLAYED APPOINTMENTS VACANT : \(£ 9.90\) per single col. inch
LINE advertisements (run-on) : 55p per line (approx. 7 words), minimum two lines.
BOX NUMBERS : 25 p extra. (Replies should be addressed to the Box number in the advertisement, c/o Wireless World, Dorset House, Stamford Street, London, S.E.1.)
PHONE: Allan Petters on 01-261 8508 or 01-928 4597.
Classified Advertisement Rates are currently zero rated for the purpose of V.A.T.

\footnotetext{
Advertisement accepted up to 12 noon Tharsday. October 4th for the November issue subject to space being available.
}

\title{
Test Engineers
}

\section*{Up to \(£ \mathbf{2 , 6 0 0}\)}

\section*{Are you seeking a challenge?}

\section*{The Post}

To meet major growth requirements \(I A L\) seek experienced Test Engineers to be engaged in testing and trouble shooting on the most advanced solid state electronic assemblies and a wide variety of sophisticated systems.

\section*{The Company}

International Aeradio Limited, a division of the British Airways Group is rapidly spreading it's interests in world markets with particular penetrations in the communications, electronics and data communications fields.

\section*{The Engineers}

Will work within the company's fast expanding electronics engineering division which is concerned with the production of solid state transmitters and receivers and digital systems equipment aimed at the Computer market.

\section*{The Requirements}

The scope and responsibilities offered within the post call for very competent engineers with proven practical experience. A technical qualification will of course be advantageous but will not be an essential requirement for the right man. Applications stating age and career to date should be addressed to: Mr. R. Radcliffe, Personnel Officer UK, International Aeradio Limited. Hayes Road, Southall. Middlesex.


\title{
SPANISH COMMUNICATIONS EQUIPMENT MANUFACTURER
}

Applications are invited from qualified design engineers specialized on:
a) Ground/Air Communications
b) TV Colour Transmitters
c) Side Band Transmitters

At least 5 years experience desirable. Company located in Madrid. Salary open.

Send resumé to:

\section*{NORTRON}

Fernando el Católico, 63
Madrid 15
SPAIN

There is scope, variety and responsibility as a

\section*{Radio Technician}

Join the National Air Traffic Services of the Civil Aviation Authority as a Radio Technician and you have the prospect of a steadily developing career in a demanding and ever expanding field.

\section*{ENTRANCE QUALIFICATIONS}

You should be 19 or over, with at least one year's practical experience in telecommunications. Preference will be given to those having ONC or qualifications in Telecommunications.
Once appointed and trained, you will be doing varied and vital work on some of the world's most advanced equipment including computers, radar and data extraction, automatic landing systems, communications and closed circuit television.

Vacancies exist at locations near London (Heathrow), London (Gatwick) and Stansted Airports and for suitably qualified people at the Signals Training Establishment, Milton Keynes, Bucks.
Salary: \(£ 1383\) (at 19) to \(£ 1836\) (at 25 or over) ; scale maximum \(£ 2158\) (higher rates at Heathrow). Some posts attract shift-duty payments. Promotion prospects are excellent and ample opportunity and assistance is given to study for higher qualifications.


\section*{SWANSEA SOUND LIMITED}

\begin{abstract}
If you believe you have suitable qualifications and/or the right practical experience, and you want to face the unusual challenge of heading up the engineering facility of the new commercial radio station in Swansea, we would like to hear from you.
Swansea Sound's new chief engineer will equip the station within the budgets and standards laid down, recruit and train the technical staff and bring the station to a fully operational state for an air-date in early summer 1974.
He will have full responsibility for his department.
The successful candidate could be a young man with a creative flair for audio electronics, or he could possess a broader experience of radio engineering but with a fresh and flexible approach to the job demanded by the particular workings of commercial radio.
In either case he will be expected to make his own contribution to the success of the station motivating and leading a compact team of technicians.
Salary will match the qualifications and abilities of the successful man.
\end{abstract}

Write in confidence to:
Keith Lunniss, Director
Radio Advertising Bureau Ltd., 35 Curzon Street

\section*{SPANISH COMMUNICAIIONS EQUIPMENT MANUFACTURER}

Has an immediate opening for An experienced Design and Development Engineer for Audio Equipment, including Highly Professional Mixing Desks, Compressors, Limiters, Audio Monitoring Amplifiers, etc. Systems Experience is desirable. Salary open.

Send resumé to:
NORTRON
Fernando el Católico, 63
Madrid 15
SPAIN

\section*{C.C.T.V. STUDIO TECHNICIAN}

Hambro Life Assurance Limited are installing a high quality, three-camera colour television studio at their offices in Old Park Lane, W.1.

Facilities will include video recording, editing and telecine. Additionally, two black and white camera and U-matic recording chains have been acquired.

We now wish to add to the small, professional team working on this project an enthusiastic, hard working individual who will be responsible for the technical quality of the studio output, keeping equipment lined up and trouble shooting where necessary. Other aspects of the job are wide and varied, including establishment of studio sets, assisting during recordings (e.g. sound or vision mixing), liaison with suppliers, etc.

Enthusiasm and experience will be regarded as more important than technical qualifications for this position. We believe that a related engineering background with an inclination to turn a hand to all parts of the production of a video-recording are the most appropriate characteristics. Filming experience, either as an amateur or professional, would be an added advantage. Salary will be negotiable around \(£ 2,500\). Working conditions are excellent and there are non-contributory pension, life assurance and B.U.P.A. schemes.
Telephone Harry Catton, 01-499 0031, for any further information and an appointment for interview.

\section*{APPOINTMENTS}

\title{
BITO OFHIERE would vau rome nshare for \(\% 2,3010\) y yar?
}

As a Radio Operator with the Post Office Maritime Service you can continue your career ashore in an interesting and expandingservice. And earn over \(£ 2,000\) a year, including compulsory pension contributions, at 25 years of age working only a 41-hour week of shift duties -with overtime this could rise to \(£ 2,300\) and possibly more.

Post Office Radio Operators benefit from a shorter pay scale than sea-going officers. You have good opportunities for promotion to positions earning basic salaries of up to \(£ 3,290\), and prospects of further advancement into Post Office Senior

Management.
To apply you need to be 21 or over and to hold a 1st class or General Certificate issued by the MPT or an equivalent certificate issued by a Commonwealth administration or the Irish Republic.

If you would like to know more, please write to the Inspector of Wireless Telegraphy, Post Office, IMTR/WTS1.1.3, Union House, St. Martin's-le-Grand, London EC1A 1AR. L52

\section*{0000000}

Foreign and Commonweath Office

\section*{Telecommunications Technicians}

\footnotetext{
at Hanslope Park, Milton Keynes, for work on various receivers and associated test equipment, recorders, telephone and teleprinter equipment, electronic ancillary equipment (some using analogue and digital tech:riques), voice frequency telegraph and other specialised equipment.

Candidates, normally aged at least 23 , must have ONC or equivalent in electrical/electronic subjects and have served an apprenticeship or had equivalent training.

Starting salary \(£ 1,628\) (at 21 ) - \(£ 1,810\) (at 23) \(£ 2.210\) (at 28 or over on entry): scale maximum \(£ 2,418\). Prospects of promotion up to \(£ 3,515\). Non-contributory pension scheme.

For full details and an application form (to be returned by 12 October 1973), write to Civil Service Commission, Alencon Link, Basingstoke, Hants RG21 1JB, or telephone BASINGSTOKE 29222 ext. 500 or LONDON 01-839 1992 (24 hour answering service). Please quote \(\mathrm{T} / 8370\)
}

The company specialise in the design and production of sound control equipment for the recording industries.

With the continued expansion of the company, vacancies arise for the following:

\section*{SENIOR TEST ENGINEER TEST ENGINEERS}

Applicants should have a theoretical knowledge of electronics and/or experience in studio engineering techniques and practices.

It would be preferable that the Senior Test Engineer has had studio experience.

Salaries by negotiation and to be commensurate with that of the position.

Apply to:

> Cadac (London) Ltd.,
> Lea Industrial Estate,
> Batford,
> HARPENDEN Herts.

Tel: Harpenden (STD 05827) 64698

\section*{The best young Engineers have computers in mind. Are you aged 21 to 25?}

Do you want a flying start to a career in computers? Here is your chance. Train as a Field Engineer with IC.L, Europe's leading computer manufacturer.
Training
You will be given thorough training on ICL electronic equipment leading to computers. Qualifications

You should be aged between 21 and 25 and be on your final year or have attained City \& Guilds electronic certificates or an HNC in electronics. You should have completed an electrical engineering apprenticeship or have at least two years' industrial experience on electronics.
Job satisfaction
As an ICL Field Engineer you have a high degree of responsibility for a customer's installation. You need technical expertise, tact and personality. So you are important as a representative of ICL.

There are opportunities of starting with us in several areas in the UK. Get the full details now by completing and returning this coupon today.

To: Mr A E Turner, International Computers Limited, 85/91 Upper Richmond Road, Putney, London SW15 2TQ.
Please send me an application form for job openings in Field Engineering.

Name
Address

ICL Computers

\section*{THE OPEN UNIVERSITY SOUND RECORDING ENGINEER}

A vacancy exists in The Open University for a sound recording engineer; five years minimum experience in the recording and editing of master tapes for radio, film, cassettes and discs. The successful candidate will be based in the sound studio (APRS member). The Open University is located in the new town of Milton Keynes and a housing scheme in conjunction with the Milton Keynes Development Corporation is operated.
Salary up to \(£ 2,241\) per annum.
Applications should be made in writing stating full details of age, experience and qualifications and names and addresses of three referees to The Acting Personnel Manager, The Open University (RE2), P.O. Box 75, Walton Hall, Militon Keynes, MK7 6AL, as soon as possible.

\author{
\([3023\)
}

\section*{BERRY'S RADIO}
has vacancies for
(a) SENIOR SALESMEN
(b) SENIOR ENGINEERS TOP RATES OF PAY

\title{
Electronics Test Engineers
}

Pye Telecommunications of Cambridge and Haverhill have immediate vacancies for Production Test Engineers. The work entails checking to an exacting specification VHF/UHF radio-telephone equipment before customer delivery: applicants must therefore have experience of fault finding and testing electronic equipment, preferably communications equipment. Formal qualifications while desirable, are not as important as practical proficiency. Armed service experience of such work would be perfectly acceptable. Pye Telecommunications is the world's largest exporter of radio-telephone equipment and is engaged in a major expansion programme designed to double present turnover during the next fiveyears. There are, therefore, excellent opportunities for promotion within the company. Pye also encourages its staff to take higher technical and professional qualifications.
These are genuine career opportunities in an expansionist company, so write or telephone without delay for an application form to:
Mrs A E Darkin at
Cambridge Works, Elizabeth Way, Cambridge CB4 1DW.
Telephone: Cambridge 51351.
or Mrs C Dawe at
Colne Valley Road, Haverhill, Suffolk.
Telephone: Haverhill 4422.

\section*{Brighton Education Committee \\ Brighton Technical College}

\section*{Senior}
C.C.T.V.

\section*{Technician} Technician required as soon as possible to
head a team responsible for the maintenance head a team responsible for the maintenance other audio visual aids.
Applicants should possess a City \& Guilds finals Appiticants should possess a City \& Guilds finals
certificate in radio and television servicing and certificate in radio and television servicin
have had relevant practical experience.
SALARY ON GRADE T4 £1.530-£1.830. (CURRENTLY UNDER REVIEW)
An additiona! allowance is payable for appropriate qualifications.

Application form and details available from:
Chief Administrative Officer,
Brighton Technical College,
Pelham Street,
Brighton, BN1 4FA.
Telephone: Brighton 685971.

\section*{ENGINEER \\ to service \\ ELECTRONIC ORGANS \\ B \& O AUDIO and C.T.V.}

The work is interesting and varied, a Company vehicle is provided and there are vacancies in Birmingham and Manchester.
Telephone or write to:
W. Swan, Jnr. or Mr. D. C. Kay,

SWAN'S
m Street, Manchester M4 ILF Tel : 061-228 3821

\section*{SERVICE ENGINEER}

LKB INSTRUMENTS LTD.
requires an additional engineer to be resident in the South London, S.E. England area for field servicing of their Scientific and Technical instruments installed in Academic, Medical and Industrial Laboratories.

The successful applicant will possess a sound basic knowledge of modern electronics and will preferably have some field experience, although this is not essential.

The Company is internationally renowned for the quality of its products and offers excellent working conditions including company car, pension scheme, superannuation and profit sharing bonus scheme.

Write for application form to
The Service Manager,
LKB Instruments Limited,
232 Addington Road,
South Croydon, Surrey, CR2 8YD [3079

\section*{Telecommunications Technicians Looking for Variety?}

The Ministry of Posts and Telecommunications needs skilled, suitably qualified Technicians to work in London on a broad range of projects. These posts offer challenge, involvement in advanced new development, and the opportunity to develop your experience across a whole spectrum of sophisticated technology
The Ministry is responsible for the planning of television and sound broadcasting, space communications, maritime and land mobile services; the technical monitoring of radio transmissions and radio regulatory control, together with the design of equipment for detecting. measuring and suppressing radio interference. There are opportunities to participate in the work of national and international conferences and there are good promotion prospects.

You should be at least 23, and must hold ONC Engineering with a pass in Electrical Engineering ' \(A\) ', or ONC in Applied Physics, or a recognised equivalent such as C\&G Telecommunications Technicians (No.49). In addition you should have at least 5 years' experience of skilled work on radio, radar or electronics. Salary: Grade II £2,593 rising to £2,902; Grade III \(£ 1,985\) at 23 £2,385 at 28 or over on entry; scale maximum £2,593. Level of appointment will depend on age and experience. Non-contributory pension scheme.
For full details and an application form (to be returned by 10 October 1973), write to Civil Service Commission, Alencon Link, Basingstoke, Hants RG21 1JB, or telephone Basingstoke 29222 ext 500 or London 01-839 1992 (24 hour answering service). Please quote \(\mathrm{T} / 8349\).


In order to keep pace with the ever increasing demand for large audio installations we require further staff.
Specifically System planners, Supervisors, Wiremen and Test Technicians.
If you consider your ability and experience suits you for one of these positions please phone or write to us.
You will enjoy long term security and involvement in extensive contracts for the manufacture of technically advanced systems for use throughout the worid.

\section*{GRAMPIAN REPRODUCERS LIMTED Hanworth Tradirg Estate,Feltham, Middlesex.}

\section*{MM PRISON AND BORSTAL SERVICE}

\section*{TRAINING INSTRUCTOR £1,853-£2,514}

LEWES, Sussex and possibly elsewhere
Inmates of prisons and borstals are given vocational training so that they may earn their living when they leave, In addition, the training is used directly in industrial work contracted for by Prison Industries.
Most of the training is to ITB specification and prepares people for City and Guilds or equivalent examinations. The service now needs the following additional instructor at HM Prison, Lewes, Sussex.

\section*{RADIO AND TELEVISION SERVICING}

To train inmates in Radio and Television Servicing and to prepare them for City and Guilds examinations.
QUALIFICATIONS Applicants should have served a full apprenticeship or have had equivalent recognised training followed by at least five years industrial experience in the radio and television and/or electronic servicing industry, City and Guilds certificate (or equivalent) is desirable. Teaching or instructing experience are added advantages.
STARTING SALARY (Civilian instructional Officer, Grade III) \(£ 1,853\) at age \(26 ; £ 2,082\) at age 28 and above, rising to \(£ 2,514\). Non-contributory pension scheme and prospects of promotion.
HOURS 40 hour week, excluding meal breaks. 4 weeks 2 days annual holidays plus \(8 \frac{1}{2}\) days public and privilege leave.
FOR AN APPLICATION FORM Please phoneor write to The Establishment Offlcer, Home Office, Port land House, Room 10/10 (17A) Stag Place, London, SW1E 5BX (telephone 01-828 9848 Ext. 666). Closing date 28 September 1973.

HOME OFFICE

\section*{Senior Quality Assurance Project Engineer}

This new appointment in the Feltham Laboratories of our Systems and Weapons Division requires a qualified engineer who is able to supervise a wide range of Quality Assurance activities concerned with
Instrumentation and Data Systems.
The duties will consist of liaison with development engineers from the project definition stage to production on Ministry Contracts.

Experience in Data Systems together with a sound knowledge of modern test equipment for functional test purposes, is essential.

Knowledge of Defence Standard 05-21/29 would be an advantage Applications should be made quoting reference number QA/ri to: Mr. J. Morrison, Personnel Officer, EMI Electronics Ltd, Victoria Road, Feltham, Middlesex
Tel. No. ox-890 3600 Extension 44

International leaders in Electionics.Records and Entertainment.

\section*{Redland Tiles Development Department at Redhill VERSATILE ELECTRONICS TECHNICIAN}

Required to construct and install a wide range of electronic and electrical control equipment using both contactor and inte. grated circuit techniques.
\(\AA\) good practical man is required capable of working on his own initiative from circuits and sketches when necessary. Good mechanica! aptitude would be an advantage.
The applicant should be qualified to drive and be prepared to travel occasionally at short notice.
May suit EXPERIENCED PROTOTYPE WIREMAN following course of instruction leading
D. F. Matthews

REDLAND TILES LIMITED
Philanthropic Road, Redhill, Surrey Telephone Redhill 64671 [3046

\section*{SOUND ENGINEER SOUTH AFRICA}

Major South African Record Company are expanding their studio operations. They require an experienced Sound Recording Engineer to head up a team that will operate a new multitrack complex with the latest equipment. Salary negotiable.
Write giving full details of professional background and experience to box WW 3067.
```

KEY SELECTION
Require for Clients
ELECTRONIC, DEV, TEST \& SALES ENGS. \& TECHNICIANS
(All grades-all areas of U.K.)
Why not give your career that much needed boost by contacting Maurice Wayne on
01.487 W411, Key Selection $01-487$ 3411, Key Selection, 126, Wigmore
Street, London, W.I.
13029

```

\section*{MARINE ELECTRONICS}

Positions available for experienced, or trainee marine electronics engineers and operator technicians. Our work involves Oceanographic Surveys and will require personel to work for periods overseas. This company is based at Yarmouth and for further information please write giving details of education and experience to:


OCEANEERING INTERNATIONAL SERVICES LTD.
Riverside Road, GORLESTON Norfolk

\section*{ilea}

\section*{Education}

Television Service
Tennyson Street, London SW8

\title{
Mobile Section Engineer
}
£2907-£3138
responsible for the technical operation and maintenance of one of the mobile control rooms, working with the Education Director and rooms, working with the Education iper with 3 a crew of two. The MCRs are equipped with 3
monochrome Plumbicon cameras, an eightmonochrome Plumbicon cameras, an eightchannel sound desk and 2 inch or 1 inch video-
tape recorders as necessary. All members of the tape recorders as necessary. All members of the
crew share rigging duties and the driving of crew share rigging duties and the driving of
vehicles. A current driving licence should be vehicles. A current driving licence should be
held and training will be given for the taking of an HGV driving test.
Applicants should possess a thorough knowledge of broadcast television engineering practices have appropriate qualifications and experience, have appropriate
Salary according to qualifications and experience.
Hours of work will be in accordance with the requirements of the service but the basic week is 35 hours. Hours are of necessity rather irregular, often involving overtime, but time off in lieu will be granted or, where that does not prove possible, overtime payment will be made. prove possible, overtime porking is very seldom necessary. The annual leave entitlement, after qualifying service, is 5 weeks and 1 day.

Application forms and details from the Education Offlcer (Ref EC/Estab 2A/2). The County
Hall, London SEe. Telh OI-633 7546 or OI633 7456. Closing date for completed applica633 7456. Closing date
tion forms October 1.


\section*{Lancashire County Council Health Department}

The Health Education Service has a vacancy for a

\section*{TECHNICIAN (TV/PHOTOGRAPHY) \\ Grade Tech. 4}

Salary \(£ 1,530-£ 1,803\)
Television is becoming an integral part of audio visual aids in the provision of health education. T.V. studio facilities are being developed and the
Health Education Service requires a technician Health Education Service requires a technician
whose duties will include the technical operation of \(T . V\). equipment.
The person appointed will, of course, be knowledgeable in the use of normal projection equip. ment. It will be an advantage for applicants to have some expertise in camera work and photography.
The post is full time, permanent, superannable and subject to medical clearance.
Application forms obtalnable from the County Medical Officer of Health, Serial No. 9693, East Cliff County Offices, Preston, to be returned by the 20th September, 1973.
[3097

\section*{MARCONI INSTRUMENTS LIMITED}

\title{
ELECTRONIC TECHNICIANS
}
are required to work on calibration, fault-finding and testing of telecommunications measuring instruments. The work is varied and will enable technicians with experience of r.f. circuits to broaden their knowledge of the latest techniques employed in the electronics and telecommunications industries by bringing them into contact with a wide range of the most advanced measuring instruments embracing all frequencies up to u.h.f.

Entrants may be graded as Test Technicians. Senior Test Technicians or Technician Engineers according to experience and qualifications. Our servicing and production programme. geared to our recognised export achievement, provides employment combined with prospects of advancement. not only within these grades. but into other technical and supervisory posts within the Company at Luton and St . Albans.

Salaries are attractive and conditions excellent. A Pension Scheme includes substantial life assurance cover provided by the Company. Assistance with removal may also be given in appropriate cases. Please write or telephone, quoting reference WW178 for application form to:

Mr. M. Leavens, Works Manager Telephone: Luton 33866. or Mr P Elsip. Personnel Officer Marconi Instruments Ltd Longacres. St. Albans. Herts
Telephone: St. Albans 59292
Member of GEC Marconi Electronics

\title{
Maintenance Engineers
}

\section*{Saudi Arabia}

Opportunities exist in Saudi Arabia for Engineers experienced in the installation, commissioning and maintenance of'complex radar systems, on-line real time computer systems and their associated peripherals, other sophisticated digital hardware, tropospheric scatter systems and telecommunication systems. These posts involve 1-2 years unaccompanied tours for which inclusive salaries between \(£ 4000\) - \(£ 6500\) will be paid. Free air-conditioned individual accommodation, and local


\section*{£4-£6500 (tax free)}
transportation is provided. U.K. leave entitlement is 16 days after each 24 weeks. Candidates, aged 23-50 years, should have extensive relevant practical experience (particularly experience of 3-D Radar), together with a formal qualification or services background. Excellent opportunities exist for career development and those selected will become permanent staff with all the benefits that this implies. Regional interviews will be conducted. Write or telephone for application form to

Chris Jamieson, Lansdowne Recruitment Limited, Design House, The Mall, London W5 5LS. Tel 01-5796585 (anytime-24 hour answering service)

\section*{TECHNICAL REPRESENTATIVES TO SELL \\ TELEQUIPMENT EDUCATIONAL PRODUCTS}

In order to satisfy the special needs of EDUCATION, three posts have been created. Successful applicants will be required to call on EDUCATION AUTHORITIES and COLLEGES in (1) England NORTH of the Wash and North Wales
(2) EASTERN counties Wash to Thames
(3) SOUTH England and South Wales

They will need sufficient knowledge of electronics to justify the product training provided by the company. Experience with oscilloscopes or knowledge of physics would be an advantage.

Basic salary \(£ 2,000\) TOTAL EARNINGS with commission and company profit share will be in the region of \(£ 2,600-£ 2,900\). Company car - Pension Scheme.
Contact Keith Retallic - SALES MANAGER
We also need TECHNICIANS at HARPENDEN, MANCHESTER and LIVINGSTONE (SCOTLAND) service centres. CONTACT MR. R. M. GARRATT

\section*{SERVICE ENGINEER EXTRAORDINARY}

We are seeking an unusual mixture to offer an exciting challenge, good prospects and pay. The service engineer sought should have both digital and analogue experience, mechanical aptitude and if possible knowledge of nuclear physics. He will cover the North of England, London and the South East, responsible for after-sales service of liquid scintillation equipment.
Please reply in confidence to:

> MANAGING DIRECTOR
> INTERTECHNIQUE LIMITED
> COTTRELL HOUSE
> 53 -63 WEMBLEY HILL ROAD,

WEMBLEY HA9 8BE requirements, we need engineers to design, supervise production, test and commission a wide range of mobile radio telephone equipment.

If you are an engineer, an ambitious service engineer or tester who is looking for a career opportunity, and you have experience on one or more of the following :-

I V.H.F. or U.H.F. radio equipment
2 Systems design
3 Circuit design of A.F., R.F., or D.C. switching circuits
4 Testing or servicing of Mobile Radio telephone equipment
5 If you can work with the minimum of supervision

Contact Mrs. L. Dyne, Personnel Dept. ERITH 3912I Burndept Electronics (E.R.) Ltd., St. Fidelis Road, ERITH, Kent DA8 IAU.

WIGGINS TEAPE RESEARCH AND DEVELOPMENT LTD.
Butlers Court, Beaconsfield, Bucks.

\section*{SENIOR} ELECTRONICS TECHNICIAN

Applications are invited for this post to lead a small team engaged in applying electronics to papermaking research and allied processes at the Central R. and D. Unit of an international papermaking group. Based at Beaconsfield the duties will include design, development, manufacture and maintenance of a wide variety of electronic, electro-mechanical and opto-electronic instrumentation.
Applicants should be of H.N.C. standard and have several years development experience with linear and digital circuits.
The salary is negotiable in the range \(\ell 2,000\) to \(£ 3,000\). The unit provides excellent working conditions, a pension scheme and luncheon vouchers.
Application forms from Mr. A. W. Massey, Personnel Department. Tel: 04945652.

\section*{CHIEF INSPECTOR}

Thorn Consumer Electronics (Chigwell) Limited is the Audio division of the Thorn Group of Companies and in order to satisfy the continuing increase in demand for our products, both at home and abroad, it has become necessary to undertake an expansion programme. A new audio factory has been established at Harold Hill in Essex, which will ultimately be the largest manufacturing unit of its kind in Europe using sophisticated production techniques.

An exceptional opportunity occurs for a suitably qualified man to join the new organisation, which will be involved in quantity volume production of high wattage unit audio equipment. as Chief Inspector.

The job will be concerned with all aspects of the inspection, test and troubleshoot functions associated with the flowline production of the units. In addition, close liaison, with the Training Department in forward pianning and training requirements will be necessary.

The successful candidate will hold suitable electronics qualifications, have experience of high volume production methods. be a capable staff motivator and will possess the drive and enthusiasm which the job will demand.

Written applications, setting out brief career details to date and current salary to:

\section*{THE PERSONNEL MANAGER, THOFIN CONSUMER ELECTRONICS, 62/70 FOWLER ROAD, HAINAULT, ILFORD, ESSEX}

\section*{LEEDS POLYTECHNIC \\ Educational Technology Unit \\ Senior \\ WorkshopTechnicianT5}
£1,803-£2,100 (under review) Ref. 13/14 This is a newly created post and the successfui candidate will be required to service and maintain electronics equipment held by teaching departments
of the Polytechnic. of the Polytechnic.
Application forms (quoting reference number)
together with further particulars, from the together with further particulars, from the
Administration Officer, Leeds Polytechnic, Calverley Administration 1513 , Leed to be returned as soon Street,
as possible.
[ 3092

\section*{CLOSED CIRCUIT TELEVISION TECHNICIAN}

\section*{\(\mathbf{f 1 , 7 1 3 - £ 2 , 7 9 0}\)}
reouired in the London Collene of Printinn .SE1, for the operation and first line maintenance of an extensive C.C.T.V. distribution system linking all the teaching areas of the college to a central control room

Duties will include the central recordina and routing of proarammes as well as the settina up of a mobile proaramme oriaination "package", consisting of cameras and associated vision and sound equipment.

Applicants should have a basic knowledde of electronics and experience with the operation or installation of broadcast or closed circuit television eauipment.

Starting salary according to oualifications and experience.
Application form returnable by October 2.
from the Establishment Officer (E/635/) Room 163 N . County Hall. London. SE 1 7PB.

Electrical Service
Engineering Group


\title{
Electronic Calibration Engineers
}

G \& E Bradley, part of the international Lucas organisation, develop and manufacture a unique range of electronic instruments and medical equipment. We can also boast the most comprehensive maintenance, calibration and repair service in the U.K.
So, to maintain our reputation and our current expansion programme, we're
looking for more experienced Electronic Calibration Engineers with a maintenance background in telecommunications, radar, microwave, ECM systems, and all types of electronic test equipment.
Professionally competent Engineers, with or without relevant qualifications, will enjoy attractive salaries
plus the benefits normally associated
with a major organisation.
And, for the ambitious, the promotional prospects to supervisory starus are exceptional.
Please apply in writing or by telephone to:
The Personnel Manager, G \& E Bradley Ltd...
Electral House, Neasden Lane, London N.W. 10.
Tel: 01-4507811

\section*{BRADLEY electronics}

A LUCAS COMPANY

\title{
Computer Engineers
}

We're going to select a number of able, decisive, productive and logical minds to become expert on our very large computer systems.

We will train you for twelve to eighteen months in basic and advanced hardware and software; after this you will be responsible for ensuring that large computer systems of a particular type are kept in first-class order. This could lead to your being in charge of a team of high-calibre engineers.

Each assignment is unique and may be in Britain, Europe or elsewhere and would appeal to the young qualified engineer who wants a high degree of responsibility, has the personality, tact and resourcefulness required of a representative of ICL and who wishes to travel whilst broadening his knowledge of large computer systems.

Applications are welcomed from experienced Computer Engineers or graduates with highly logical minds and a degree in maths, physics or engineering science.

You will initially, and during training, be based in one of the following locations:

Letchworth/Stevenage
West Gorton, Manchester
Kidsgrove, Staffordshire
Write for an application form, quoting reference WW 496 C , to A E Turner, International Computers Limited, 85/91 Upper Richmond Road, Putney, London \(\mathrm{SW}_{15} 2 \mathrm{TQ}\).

We have vacancies for:

\section*{SERVICE TECHNICIANS}
based at Camberley, to work in a specialised Department dealing with miniaturised transmitter/receivers on fault diagnosis and correction. Technical experience of communications systems is an advantage.

\section*{FIELD SERVICE ENGINEERS}
in the Greater London Area. Applicants should have experience in fault finding and testing of UHF/VHF radio equipment. Current driving licence essential, company vehicle provided.
The Company has much to offer those who are interested in the sophisticated, modern world of radiotelecommunications and who can demonstrate their ability in this field.
Please contact:

\section*{The Personnel Officer, STORNO LTD.,}

Frimley Road, Camberley, Surrey.

\section*{ELECTRONIC ENGINEERS}
required for equipment maintenance and associated engineering projects. Knowledge of professional tape recording equipment, studio operations, or high speed tape duplicating systems is desirable. Salary will be according to age and experience. Please write giving details of age, qualifications, experience and present salary to Chief Engineer, Rediffusion Reditune Ltd., Cray Avenue, Orpington, Kent.


REDIFFUSION

\section*{SUPERVISOR}
(EVENING SHIFT)
A music cassette/cartridge plant requires a man aged 28 or over to supervise their duplicating department. Applicants must have some practical experience in the audio field, and an understanding of tape recording principles is essential.
We can offer good working conditions and a salary in the region of \(\mathbb{I} 750\) p.a. dependant on ability and experience.

Apply to: Mr. R. A. Goodwin
Trident Tape Services,
4/10, North Road
Tel: No. 6090087.
[3058

UNIVERSITY OF KENT at Canterbury A
TECHNICIAN (Grade 3)
is required for the Audio Visual Aids Service He must be experienced in present-day television servicing and, desirably, should also have a keen interest in cine-projection and other visual aid to teaching. The salary scale is E1539-E1794. Further particulars and application forms may be obtained from R. Robson, Assistant Registrar quoting reference T73/10. The closing date for completed applications is 29 th September, 1973

\section*{Ulster-The New University INSTITUTE OF \\ CONTINUING EDUCATION \\ MAGEE UNIVERSITY COLLEGE LONDONDERRY \\ C.C.T.V. TECHNICIAN}

Applications are invited for the above post. Duties will include the operation and maintenance of CCTV services and preparation of programme material.
Qualifications: HNC, or equivalent, plus at least seven years previous experience.
Salary scale: \(£ 1,881-£ 2,241\) per annum.
Application forms and further particulars should be obtained from The Registrar, The New University of Ulster, Coleraine, Co. Londonderry Northern Ireland (quoting Ref. No. 73/130/149/ \(31 / 98\) ) to whom completed applications, including the names and addresses of three referees, should
be returned not later than 31 St October, 1973.

\section*{UNIVERSITY OF SHEFFIELD AUDIO VISUAL TECHNICIAN}
(Grade 3) responsible for maintenance and operation of audio visual equipment (film, slide and overhead projectors, tape recorders. P.A. systems, photo-copying, etc.).

Familiarity with photographic techniques an advantage. Salary scale \(£ 1,539 \times 651\) (5)E1,794 p.a.

Write giving details of experience and qualifications. to the Deputy Director of Services (Ref. B.333/BW), The University. Sheffield, S10 2 TN .

\section*{KING'S COLLEGE HOSPITAL} MEDICAL SCHOOL
(Unlversity of London)
Denmark Hill, London SE5 8RK

\section*{ELECTRONICS \\ EXPERIMENTAL OFFICER}

A vacancy exists in the Department of Biomedical Engineering for an Experimental Officer to work as part of a multi-disciplinary team on the development and construction of prototype electronic instruments for use in medical research. Salary will be in the range of exte the ind according to age and experience and the appointment will be for two years in the first instance. Candidates should have had adequate experience
either in industry or in hospital and will be exeither in industry or HNC in electronics or light pected to hold an HNC in electronics or
current electronic engine Applications to the Director, Department of Biomedical Enginnering.

\section*{ELECTRONIC ENGINEER}

Recording Studio of major Record Company require young Audio Electronic Engineer with a fresh approach to the problems of modern recording electronics.

POLYDOR RECORDS
STUDIO LONDON
01-499 8686 ext 51

\section*{IPSWICH AND DISTRICT}

HOSPITAL MANAGEMENT COMMITTEE

\section*{ELECTRONICS TECHNICIAN}

Applications are invited for the above post. Candidates should possess H.N.C. or equivalent qualifications qualified and experienced candidates to suitably
in this field.

SALARY SCALE: \(\mathbf{6 1 , 6 0 2}\) to \(\mathbf{6 2 , 0 7 6}\) p.a. According to qualifications and experience.
Please note: that although the salary scale for this post rises from \(£ 1,602\) to 62,076 . new entrants to the Health Service are normally required to start on the lowest point of the scale. The successful candidate will be a member of a new and expanding department. servicing a

For further details of this post, please telephone lpswich 5648 I . Ext. 33 or write to:- The Group Engineer, Ipswich and Distriet Hospital Management Committee. 26 Broughton Road, IPSWICH, IPI 3QS.

\section*{Draughtsmen!}

EMI, one of the world's leading Companies in the fields of military and commercial Electronics, are forming teams to undertake several new and important projects in their divisions at Hayes, Middlesex challenging work on the design of highly advanced electronic equipment. This opens up first class opportunities for

\section*{Senior} Intermediate and Junior Draughtsmen. with a mechanical,electrical orelectronics background.

To work in the following fields
* Radar and associated projects
* Television and Aerial Systems
* Microelectronics - thick and thin film work
* Microwaves tubes and components

Starting salarics up to \(£ 2600\) p.a. If you are looking for a good career in the forefront of Electronics technology, please write giving carcer details, or phone for an application form: Richard Black, Personnel Department, EMI Electronics Ltd., 135 Blyth Road, Hayes, Middlesex.
Tel: OI-573 3888 Ext. 2887.

\section*{METROPOLITAN WATER BOARD}

\section*{Assistant Communications Officer \\ (£2,733-£3,105)}

Applications are invited for the above-mentioned post which will be based at the Bcard's Head Offices.
Duties are to assist in project investigation, planning and development of mobile radio, telecommunications (telephone and telegraph) and data transmission networks.
The successful candidate should possess a suitable qualification, e.g. Degree, H.N.C., etc., and be experienced in Post Office procedures and have had practical experience at supervisory level in one of the following fields for at least five years.
(1) V.H.F./U.H.F. mobile radio communications.
(2) Telecommunications-privately switched networks, large P.A.B.X. installations (cross-bar type), transit switching techniques, Post Office telephone equipment, and data transmission.
A.n application form returnable by 5 th October, 1973, may be obtained from the Establishment Officer, Metropolitan Water Board, New River Head, Rosebery Avenue, ECIR 4TP, or telephone 01-837 3300 Ext. 19.

\title{
Telecommunications Engineer
}

\author{
With the expansion of the company's off-shore oil and gas production activities in UK and overseas areas, a further telecommunications engineer is required to work in our London Office \\ Duties will include the design, development and implementation of remote control and supervisory systems with microwave/tropospheric scatter or cable links and telecommunications systems for normal correspondence and data. Some overseas travel will be involved for site surveys and equipment installation/commissioning \\ Candidates, aged \(30-40\), must be qualified C. Eng. or B.Sc. in Electronic Engineering. Experience to include at least four years in a senior capacity on installation, testing or site maintenance of electronic telecommunications equipment and four years design and development in systems planning. At least one year's service in the oil or allied industries in this specialised field is desirable. \\ A realistic salary will be paid, with excellent terms and conditions \\ \(\square\) Please write in the first instance, giving details of your experience, quoting reference AAG.604, to: The Manager, Central Recruitment. The British Petroleum Company Limited. Britannic House, Moor Lane, London, EC2Y \(98 U\).
}

\section*{DYNAMIC SERVICE ENGINEERS DIGITAL ELECTRONICS}

OHIO-NUCLEAR is the world's leading supplier of radio-isotope imaging equipment for use in diagnostic nuclear medicine. Due to continued rapid expansion we wish to recruit experienced field service personnel to undergo specialised technical training on our advanced systems to become PRODUCT SPECIALISTS (SER VICE).

The work involves travel to install and service our range of digitally-oriented medical equipment throughout the U.K. (occasionally abroad).

DESIRABLE REQUIREMENTS for these positions are as follows:
- H.N.C. or equivalent
- wide, general electronics background
- thorough knowledge of 74 Series logic
- working knowledge of core and tape memory systems
- ability to work on own initiative with minimum supervision.
These are KEY POSITIONS offering real scope for advancement in a very exciting and rewarding field. The posts command an attractive salary, company car and extensive fringe benefits.
Detailed résumé, including age, education and experience, to:

The Service Manager, Ohio-Nuclear (U.K.)
Radix House, Central Trading Estate,
Staines, Middlesex.

\section*{SPANISH \\ FIRM \\ NEAR MADRID}
is looking for design and development engineers with a minimum of three years of experience in the field of P.C.M. equipment to be used by the telephone industry.
Areas of interest are encoders and decoders. P.C.M. multiplexers and R.F. equipment to transmit P.C.M. data.
Salary open.

Send résumé to:
NORTRON
Fernando el Católico, 63
Madrid 15
SPAIN
2584

\section*{HARTLEPOOL HOSPITAL MANAGEMENT COMMITTEE}

\section*{Electronics Technician}

Applications are invited for this new post to undertake the maintenance of electronic equipment in hospitals in and around Hantlepool.

The appointment is to the Staff of the Group Engineer, to whom the successful applicant will be accountable for the repair and planned preventive maintenance of a wide range of electronic equipment, including electro-medical and laboratory apparatus, H.F. pocket-paging receivers and itransmitters, engineering controls, and audio/radio frequency distribution systems.

Applicants should preferably hold H.N.C. (Electronics or light current Electrical engineering) or City and Guilds Full Telecommunications Certificate.

National Health Service Conditions.
Starting Salary within the scale of \(£ 1,602-£ 2,076\) p.a., depending on age, experience and qualifications.

Application forms are available from:-
GROUP PERSONNEL OFFICER,
hartlepool hospital management committee, CENTRAL ADMINISTRATION,
GENERAL HOSPITAL,
hartlepool.

\section*{THE HOSPITAL FOR SICK CHILDREN} GREAT ORMOND STREET, LONDON, WCIM 3JH

\section*{ELECTRONICS TECHNICIAN}
required by the Medical Electronics Workshop to set and maintain high standards of serviceability and safety of a wide range of electronic apparatus used in the workshop and the Hospital.
Applicants must have sufficient constructional ability to enable them to assist with the modification and improvement of both existing and new apparatus. Experience of medical equipment would be an advantage but is not essential.

Salary on a fixed scale of \(£ 1,728\) to \(£ 2,202\).
Qualifications should be a minimum of O.N.C. or equivalent.

Day release for further study is available.
Application forms may be obtained from the Deputy Hospital Secretary.

\section*{AUDIO SERVICE ENGINEERS}

Lindair, London's leading Hi-Fi Retailers, require 2 Audio Service Engineers for their Service Dept. in the West End.

The Service Department deals with leading top-quality brands of \(\mathrm{Hi}-\mathrm{Fi}\) equipment, such as Amps., Receivers, Tape-Recorders, and Loudspeakers.

Applicants must be fully competent to work under the minimum of supervision. Ideally they should have appropriate City and Guild and/or " O" levels in Maths and Physics, but qualifications are not as important as the ability to complete a job once started.

Minimum age 22.
High Salaries will be paid, other attractive conditions of service include 3 weeks' paid holidays.

To arrange an interview write or phone:-
Mr. George Welsh, Lindair, Kirkham House,
54a Tottenham Court Road. London. W1
Telephone: 01-6371601
LINDAIR

\section*{RADIO OFFICERS}

\section*{DO YOU HAVE PMG I PMG II MPT 2 YEARS OPERATING EXPERIENCE}

POSSESSION OF ONE OF THESE QUALIFIES
YOU FOR CONSIDERATION FOR A RADIO
OFFICER POST WITH COMPOSITE SIGNALS
ORGANISATION.
On satisfactory completion of a 7 -month specialist training course, successful applicants are paid on a scale rising to \(£ 2.527\) pa: commencing salary according to age - 25 years and over \(£ 1,807\) pa. During training salary also by age, 25 and over \(£ 1.350\) pa with free accommodation.

The future holds good opportunities for established status, service overseas and promotion.

Training courses commence at intervals throughout the year. Earliest possible application advised.

Applications only from British-born UK residents up to 35 years of age ( 40 years if exceptionally well qualified) will be considered.

\section*{Full details from}

Recruitment Officer, Government Communications
Headquarters, Room A/1105 Priors Road, Oakley, Cheltenham, Glos GL52 5AJ, Telephone: Cheltenham 21491 Ext 2270

\section*{ARPOINTMENTS}

\section*{CENTRALELECTRICITY GENERATING BOARD SOUTH WESTERN REGION}

\section*{SCIENTIFIC SERVICES DEPARTMENT CONTROL \& INSTRUMENTATION BRANCH}

\section*{LABORATORY TECHNICIAN}

A vacancy exists in the Instruments Section of the Scientific Services Department at Portishead. Somerset.
The work will be largely based at the laboratories at Portishead but site work at the Region's Power Stations will be necessary.
The successful candidate will join a new group. in a supporting role, to Research Officer involved in using mini computers and other logic devices to control, collect and analyse data from experiments. The group also provides a service in design and constructing electronic signal conditioning units.
The candidate's duties will be to construct electronic equipment and to generally assist in the production of both hardware and software associated with computer control and data collection. He will be expected to undertake some electronics development work.
Applicants should have an Ordinary National Certificate in Electrical/Electronic Engineering or Physics, and some practical experience in electronics and circuit construction. A working knowledge of digital and printed circuit techniques and workshop practices is desirable.
The terms and conditions of service will be in the National Joint Board Agreement for the Electricity Supply Industry. The salary will be within the range \(£ 1.446\) to \(£ 2.376\) per annum depending upon age. experience etc. In addition, allowances of \(£ 60\) and £ 174 per annum are payable.
Applications on Form SF/1, obtainable from the Personnel Manager, 15-23 Oakfield Grove, Clifton, Bristol BS8 2AS, should be returned to him (quoting Vacancy No. 292/73) by not later than 1 October 1973.

\section*{Are you interested in}

\section*{Communal Aerial Television Systems Work?} Then read on further......

Due to continued expansion, EMI Service, part of EMI's Electronics and Industrial Operations group of Companies, has the following vacancies for engineers at Hayes, Middlesex.

\section*{SERVICE ENGINEERS}
required for bench and field work on Communal Television Aerial equipment. Must be capable of diagnosing faults and repairing wide range of aerial amplifying and distribution equipment.

\section*{SYSTEMS PLANNING}

ENGINEERS
for the planning of Communal Television Aerial installations. Previous experience required to be capable of producing practical plans from building details and subsequently setting to work after installation.

Attractive starting salaries. Contributory Pension Scheme. Assistance with removal expenses in appropriate cases.

\section*{WANT TO TAKE THINGS}

FURTHER
then write or telephone for an application form to:
R. N. L. Black, Personnel Department, EMI Limited, 135 Blyth Road, Hayes, Middlesex. or-573 3888, Ext 2887.

The Department is engaged upon providing diagnostic services, using radioactive isotopes including scanning, ultrasonic, electromyography, patient monitoring, etc. Apossess an ONC, two qualified Radiographers or science subject or " \(A\) " levels in an appropriate educational qualifications will be braing for higher appropriate cases.
Salary \(\subset 1.209\) rising by annual increments to a maximum of \(\{1.563\) per annum.
Applications giving full details of education, qualifications and experience together with the names of two referees to the Hospital Secretary. St. George's Hospital, Lincoln.
[3028

\begin{abstract}
DO you require Indian representation? B.Sc. bay. Willing to act as local fep for U.K. companies Ref available. Write Box WW 3035. ELECTRONICS TECHNICIANS. (1) Grade II E \(£ 2,037-£ 2,634\). To be accountable to the Group Engineer for the Department. (2) Grade III \(£ 1,725\) £2,202. Salaries under review for payment of Government Phase II increase. The successful applicants will service and calibrate a wide range of equipment used in medical. surgical and engineering services, working closely with medical and other may be available. Further particulars and application forms returnable by October 8th from Mrs. J. Moore, Staffing Officer, Ext. 2202. Northwick Park Hospital and Clinical Research Centre, Watford Road, Harrow Middlesex, HA1 3UJ. Telephone 01-864 5311. [3020 FULLY experienced Audio Engineer required by T leading importer of high fidelity equipment. The applicant should have working experience with most types of unit and be fully up-to-date with current techniques. 5 day week. Salary by negotiation Applications to the Managing Director, HowlandWest Ltd.. 3-5 Eden Grove. London N7 8EQ. 13062 HI-FI AUDIO ENGINEERS. We require experito get them. Tell us about your abilities. \(01-4374607\).
\end{abstract}

MPERIAL College of Science and Technology Tech1 nical Vacancy ELECTRONICS DEVELOPMENT ENGINEER required, for a research project requiring development of electrical controls for high pressure hydraulic and mechanical systems. Qualifications: Degree or equivalent, mechanical background. Salary £2,229 to \(£ 2,715\) plus \(£ 175\) per annum London weighting, according to experience. Contract for a limited period. Application forms from Departmental Superintendent, Department of Geology, Imperial College, London SW7 2BP. [3044 LEEDS (ST. JAMES'S) UNIVERSITY HOSPITAL LHYSICS TECHNICIAN COMMITTEE - MEDICAL PHYSICS TECHNICIAN (GRADE IID) (NEW POST). An electronics technician is required for the maintenance of x-ray image intensifiers and closed circuit in Leeds. There are now 14 sets of equipment in use. The person appointed will work with the x-ray maintenance staff in the Medical Physics Unit. Candidates should have an O.N.C. or H.N.C. in electronic engineering or a science degree, followed by at least three years of relevant experience. Salary scale \(£ 1,602-\) £2,076. Application forms available from the Group Personnel Manager, St. James's Hospital, Leeds LS9 7TF. Closing date September 28th, 1973. [3045 'TEST Engineers and Installation Engincers-starting for bath \(£ 1,400-\mathrm{t} 2,000\). Electrosonic have vacancres visual and lighting control equipment Permanent and pensionable positions offering an excellent opportunity for applicants with initiative, experience and a sound knowledge of electronics. Starting salary according to experience. Electrosonic is an expanding international company and posts offer opportunities for travel. Apply Electrosonic Ltd., 815 Woolwich Road, Charlton, SE7 8LT. Telephone 01-855 1101 between hrirs \({ }^{\text {CECHNT }}\) am. 3087 TECHNICIAN. GRADE 5 required for Electronics 1 Section of Science Workshop. Applicants should be suitably qualified and be conversant with the maintenance and adaptation of electronic instruments and equipment and the design and construction of 5 simple electronic apparatus for research and teaching. Commencing salary on scale \(£ 1.881 \times 72\) Scheme plus \(£ 175\) London Allowance, according to qualifications and experience. Application forms may be obtained from The Assistant Secretary (Personnel), (WW) Bedford College, Regent's Park, London. NW1 4 NS. (Tel.: 01-486 4400 Ext. 313). [3060 CLASSIFIEDS-Continued on p. 125

\section*{APPOINTMENTS}

\section*{SALES MANAGER}
to market commercial sound products including the newly acquired 'Altec' Agency.
Salary negotiable: \(£ 3,000\) plus.
Apply in writing to:
theatre projects sound limited,
10 Long Acre, London, W.C. 2
[3112

\section*{Electronics Engineer}

We are a London-based Studio and we are looking for a young Engineer who is capable and eager to learn. He will be working with Broadcast Colour cameras, VTR and Telecine.
Write or phone: Jon Hocking, Technical Director,
zOOM TELEVISION LIMITED 15-19 New Fetter Lane, London E.C. 4
Tel. 01-353 3641

London Borough of Haringey Education Service

\section*{Laboraiory Technician}

Salary \(£ 1,416-£ 1,635\) per annum plus recent pay award. Commencing salary according to qualifications. Laboratory Technician required at Full-time Laboratory Technician required at
Stationers' Company's 5chool, Mayfield Road, N.8, to work 36 hours per week \(\times 52\) weeks per
annum.
Minimum qualifications: Ordinary National Certificate or Ordinary National Diploma; City and Guilds Laboratory Technicians Certificate; 4 G.C.E. passes with 2 at ' \(A\) ' Level in appropriate subjects; Membership of Institute of Science Technology OR an equivalent suitable qualification OR 5 years suitable experience. Qualifications in Electronics would be an advanta.ge.
Candidates will be responsible for the maintenance
of the Language Laboratory and will be required of the Language Laboratory and will be required throughout the School and help monitor a computer link-line.
The post is ideal for a candidate who wishes to gain experience in the maintenance of a fairly wide range of equipment.
Application forms obrainable from Chief Education Officer, Somerset Road, N.17, returnable by 29 October, 1973.

\section*{CLASSIFIEDS-Continued from p. 124}

\section*{MARINE RADIO OFFICER: (Aged 27), eight 1 years sea service seeks employment ashore within the electronics field in either field service, sales engineering or communications. PMG Certificate (2), DTI Radar Certificate. Willing to learn new fields and travel either at home or abroad. Permanent position, preferably in Scotland. Box WW 3099. \\ SURPLUS components and computer boards pur- \\ ARTICLESTFOR SALE}

A MPEX FR 1400 Instrumentation Tape Recorder A Reproducer, £375.00. EMI TR 52 two-channel audio Tape Recorder, £120.00. EMI magnetic drum Echo System (ex Abbey Road), £185.00. Hewlett Kearfott Gyro T2502, £7.50. ERA solid-state 400 Hz Inverter, \(100 \mathrm{VA}, £ 15.00\). Tinsley 50 Hz Tuning Fork int polished wood case, \(£ 12.00\). IBM Standard Electric Typewriter mounted on table, \(£ 55.00\). IBM Selectric Typewriter on table, £95.00. Solartron CD 1400 d.b. Oscilloscope, \(£ 140,00\). Solartron LM 903 a.c. Converter/Voltmeter, \(£ 12.00\). ICT adjustable 6 V Suppiy with twin meters, f16.00. Deltron \(12 \mathrm{~V} / 20 \mathrm{~A}\) digit ttl Counter-Timer, £25.00. Ekco M 5024 six-
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\section*{Semiconductors: Basic Theory and Devices}

\author{
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\(60 / 39.7 / 0.3\)
\(50 / 50 \mathrm{Sn} / \mathrm{Pb}\)
\(50 / 49.7 / 0.3 \mathrm{Sn} / \mathrm{Pb} / \mathrm{Sb}\)
\(50 / 48.5 / 1.5 \mathrm{Sn} / \mathrm{Pb} / \mathrm{Cu}\)
\(45 / 55 \mathrm{Sn} / \mathrm{Pb}\)
\(40 / 60 \mathrm{Sn} / \mathrm{Pb}\)
\(40 / 59.7 / 0.3 \mathrm{Sn} / \mathrm{Pb} / \mathrm{Sb}\)
\(30 / 70 \mathrm{Sn} / \mathrm{Pb}\)
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\section*{(nominal major elements)}

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63/36.7/0.3 Sn/Pb/Sb
\(60 / 40 \mathrm{Sn} / \mathrm{Pb}\)
39.7
\(50 / 50 \mathrm{Sn} / \mathrm{Pb}\)
-
\(\mathrm{Sn} / \mathrm{Pb} / \mathrm{Cu}\)

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\hline \multirow[b]{2}{*}{Grade} & \multicolumn{3}{|l|}{Melting Temperature} \\
\hline & & & Specification \\
\hline TI.C & 145 & 145 & DIN 1707 \\
\hline LMP & 179 & 179 & DIN 1707 \\
\hline Sn62 & 179 & 179 & QQ-S-571E \\
\hline Sn63 & 183 & 183 & QQ-S-57 1E \\
\hline K & 183 & 188 & H.S. 219 \\
\hline Sn60 & 183 & 188 & QQ-S-57 1E \\
\hline F & 183 & 212 & B.S. 219 \\
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DTD 900/4535 \\
DIN 1707
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\hline R & 183 & 224 & B.S. 219 \\
\hline G & 183 & 234 & B.S. 219 \\
\hline Sn40 & 183 & 234 & QQ-S-571F. \\
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