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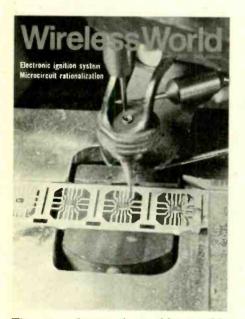
Wireless World

Electronics, Television, Radio, Audio

Fifty-ninth year of publication

January 1970

Volume 77 Number 1411



The macrophotograph on this month's cover shows wires being bonded on to integrated circuits at the Mullard Southampton works. On page 6 the future of linear i.cs is discussed.

OUR NEXT ISSUE

Loudspeaker performance: Paul Klipsch, originator of the Klipsch horn, compares horns and direct radiators

Ceramic pickups and transistor pre-amplifiers: are they incompatible? Matching: what is meant by this term?



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Contents

- 1 Editorial Comment
- 2 Capacitor-discharge Ignition System by R. M. Marston
- 6 The Future of Linear I.Cs by R. Hirst
- 8 News of the Month
- 10 Letters to the Editor
- 11 Circuit Ideas
- 13 Amorphous Semiconductors by J. E. Carroll
- 16 Application Notes
- 17 Magnetoresistance by B. E. Jones
- 19 Announcements
- 20 Low-distortion Bias and Erase Oscillator by D. Griffiths
- 23 Industrial Telemetry by R. E. Young
- 27 Active Filters 6 by F. E. J. Girling and E. F. Good
- 32 Personalities
- 33 Progress in Tape-recording Techniques by S. Feldman
- 34 World of Amateur Radio
- 35 Electronic Metronome by D. T. Smith
- 35 A Digital Christmas Tree
- 36 D.C. Bias in Push-pull Power Amplifiers by R. A. Smith
- 38 1970 U.K. Conferences & Exhibitions
- 39 New Products
- 44 Meetings
- 45 Test Your Knowledge questions and answers devised by L. Ibbotson
- 46 Literature Received
- 46 H.F. Predictions
- 48 Real & Imaginary by "Vector"
- A95 SITUATIONS VACANT
- A114 INDEX TO ADVERTISERS

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Wireless World

"Explosion in communications"

This was the expression used by Mr. Stonehouse, Minister of Posts and Telecommunications, when he announced in the House of Commons that he was considering setting up an independent enquiry into the long-term future of broadcasting in the U.K. He referred to the need to examine the implications of new technologies, e.g. the possibility of 100 or more communication channels going into every home in Britain via a single wire or microwave link, "which would bring about an explosion in telecommunications".

One's immediate reaction to the proposed broadcasting enquiry is "what, another one!" It will be recalled that very few of the recommendations in the Pilkington report of 1962 were implemented and over the years there have been many proposals made by the various committees of enquiry or commissions which, maybe because they were too sweeping, have been turned down. It would, however, appear from the Minister's latest statement that if the proposed committee is set up it will be asked to look at the long-term future of internal telecommunications generally and not just broadcasting. If so, this is going to be a gargantuan task calling for technological forecasting. Incidentally, Professor W. H. G. Armytage, of the University of Sheffield, speaking recently to members of the Institution of Mechanical Engineers, said it now seems that "technological forecasting is, like weather forecasting, very respectable". He pointed out, however, that technological forecasting must not be confused with "the inspired doodling that has characterized science and engineering through its history . . . nor the intuitive forecasts that enabled writers like . . . Hugo Gernsback to predict radar or Arthur C. Clarke to predict the earth satellite". Professor Armytage defined it as "the application of scientific method-or objective, almost clinical method-to the analysis and forecasting of technological change"

Bearing this in mind the proposed committee could produce a really far-seeing forecast of telecommunications into the '80s and beyond; but the members will need to be supermen or they will find themselves bogged down by tradition and vested interests.

What are the prospects? The idea of a super telecommunications grid covering the whole country has frequently been suggested and with the growing use of telemetry and control systems, as evidenced by a contribution in this issue, it is fast becoming a necessity.

Such a super grid will be an amalgam of radio and cable techniques. Despite our title we are not so bigoted as to be blind to the potentialities of cable for distribution networks. It is, however, worth recalling that in the early days of this journal there was a fierce war waged between cable and wireless (apocryphal stories are told of the sabotage of cable systems in order to show that wireless was inviolable!) but a marriage was arranged. It is, of course, true to say that without radio devices (amplifiers, repeaters, and the like) the present cable networks could not have materialized.

We do not intend to gaze into our crystal ball, engage in inspired doodling or make intuitive forecasts, but the future for the electronics and radio engineers is certainly exciting.

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Capacitor-discharge Ignition System

An electronic ignition system, suitable for any car, which offers a large number of advantages over conventional ignition

by R. M. Marston

2

When this unit is wired up to a car's existing ignition system it greatly improves the shape of the ignition voltage waveforms, and enables a more stable flame-front to be generated in the engine's cylinders. Better combustion is thus obtained, and engine performance is considerably improved.

The unit, which is known as a capacitor-discharge ignition system, confers an impressive list of benefits in terms of engine performance. It gives easy starting, even under sub-zero conditions, and also gives immunity to performance deterioration due to contact-breaker bounce. In addition it gives quicker engine warm-up, improved acceleration, better high-speed performance, and improved fuel economy (2 - 5%). Even more important, it virtually eliminates contact-breaker point burning and wear, gives greatly improved spark-plug life (typically 3 to 5 times longer than in conventional ignition systems), and overcomes the need to adjust contact-breaker and spark-plug gaps with precision.

The ignition unit can be added to any car fitted with a conventional 12-V coil ignition system irrespective of the number of engine cylinders.

Fig. 1 shows the circuit of a conventional, or inductivedischarge ignition system. The contact-breaker (c.b.) points are opened and closed by an engine-driven cam. When the points are closed, current from the battery builds up in the coil primary, to a maximum value of about 4.5A exponentially, with a time constant of L/R seconds, typical time constants are between 2 and 10ms. As the current builds up, it stores an energy 'packet' of $(L.I^2)/2$ joules, or wattseconds, in the coil primary.

When the points open, the primary current collapses rapidly via C_1 , and induces a peak potential of about 300V across the coil primary; this voltage is increased to about 30kV at the secondary winding, and this energy is transferred to the spark-plugs by the vehicles distributor. C_1 and the coil form a resonant circuit when the points are open, and the secondary voltage takes about 125μ s to build up to its peak value.

Fig. 2 shows typical inductive discharge ignition performance characteristics and ignition requirements at different engine speeds; the early part of the graph, up to about 100 r.p.m., indicates typical sub-zero starting conditions, when battery voltage falls to about 10 V, compared to a normal value of 13.5 V when under dynamo charge. Note that the system operates with very little safety margin under coldstart conditions, and that the available secondary energy becomes inadequate when engine speeds reach 5,900 r.p.m., so that misfiring starts to occur above this speed.

Finally, the relatively long secondary voltage rise times of the inductive system (typically about 125μ s) make the ignition system very vulnerable to high energy losses due to fouling of the spark-plug gaps by carbon and oil deposits. These deposits act as a resistance (typically about $2M \Omega$ in cases of bad fouling) across the points. These deposits inevitably absorb some of the applied energy (power-time), and total energy absorption increases in proportion to voltage rise time and fouling resistance.

Capacitor-discharge ignition systems, on the other hand, suffer from hardly any of the snags outlined above. Fig. 3 shows the block diagram of the particular ignition system described here. A self-regulating voltage converter is used to charge storage capacitor C_1 to 400 V, almost irrespective of actual battery potential. When fully charged, this capacitor stores 0.08 joule.

When the c.b. is closed, zero input is applied to the pulse shaper, and the thyristor is off; a standing current of about 250mA is passed through the c.b. via R_1 under this condition, to keep the points 'clean'. The converter is operating, and charges C_1 to 400 V; the capacitor has a charging time constant of about 1.6ms.

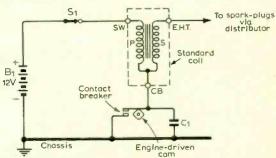


Fig. 1. Circuit of a conventional inductive discharge ignition system.

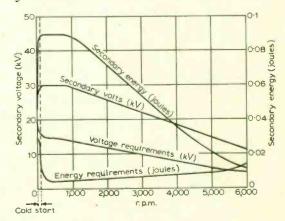


Fig. 2. Typical performance of the circuit of Fig. 1 together with engine energy requirements. The curves assume that battery voltage is normally 13.5V falling to 10V at cold start.

Wireless World, January 1970

When the c.b. points first open, the pulse shaper operates and turns the thyristor on in about 2µs. This short circuits the output of the converter, and turns it off. Simultaneously, one side of C_1 is connected to ground and discharges rapidly into the primary of the coil; the coil steps the resulting primary voltage up to about 40kV, and the stored energy of C_1 is transferred to the spark-plugs. The secondary voltage has a rise time of only a few microseconds. C_1 and the coil form a resonant circuit when the thyristor is on, and have a typical resonant frequency of 1600Hz, giving a period of roughly 600µs. At the instant the thyristor fires, the coil's primary voltage rises (in about 2μ s) to 400 V, but 300 μ s later the voltage falls to zero as the circuit oscillates and the thyristor turns off, preventing further oscillation. Once this happens the voltage converter re-starts and begins to re-charge C_1 , even though the c.b. points may still be open. The process is repeated when the c.b. opens initially again. Note that the primary coil voltage is isolated from the vehicle's c.b. terminals, which are thus subjected only to the moderately low voltages and currents.

Fig. 4 (a) shows the actual spark voltage performance of

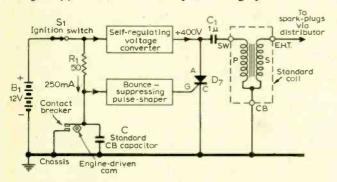


Fig. 3. Block diagram of the capacitive discharge ignition system described in the article.

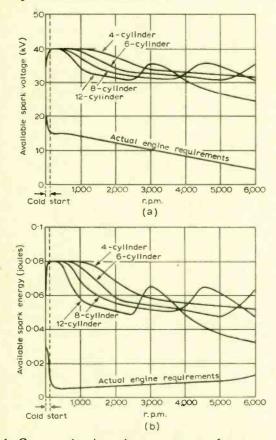


Fig. 4. Curves showing the measure performance of the capacitive discharge ignition system with the same battery voltage conditions as Fig. 2.

the prototype system at different engine speeds, when fitted to different types of engine, together with worst-case ignition voltage requirements, and Fig. 4 (b) shows the system's energy generating performance. Note that both the available voltage and energy are well in excess of engine needs under all operating conditions.

3

The full circuit of the negative ground version of the ignition system is shown in Fig. 5. C_1a and C_1b form the 1μ F energy storage capacitor. Tr_1 - Tr_2 , T_1 and the D_3 - D_6 bridge form the self-regulating voltage converter. Tr_3 and its associated network form the bounce-suppressing pulse shaper, which fires the thyristor via C_3 .

The voltage converter section operates as follows: Tr_1 , and Tr_2 are an astable multivibrator which uses the halves of the centre-tapped primary of T_1 as collector loads and which generates a series of 24 V (approximately) square waves at each collector, at a frequency of roughly 50Hz. The inductive nature of T_1 causes the early part of each square wave to shoot above the normal flat top; R_{11} - R_{12} and zener diodes ZD_1 and ZD_2 are used to limit this overshoot to 28 V peak. T_1 steps the square waves up to 400 V peak at the secondary winding. This voltage is then converted to d.c. via the D_3 - D_6 bridge rectifier, and used to charge C_1 . It is this overshoot regulation that gives the ignition system its good cold-starting characteristics. R₆ gives the circuit a degree of protection in the event of the battery voltage (under dynamo charge) rising above 15 V, and at the same time reduces the C_1 voltage at high engine speeds.

It should be noted that, although the converter oscillates at a natural frequency of only 50Hz, it is in fact capable of giving good spark generation at c.b. frequencies in excess of 660Hz, i.e., above 20,000 r.p.m. in a four-cylinder, and above 10,000 r.p.m. in an eight-cylinder engine.

At the moment that the c.b. points first open in each ignition cycle the thyristor is triggered, so Tr_1 and Tr_2 stop oscillating; 300μ s later, the thyristor then turns off, so the multivibrator starts oscillating again. The start of the first half cycle of each converter operation is thus synchronized by the c.b. At c.b. frequencies above about 100Hz, therefore, the converter starts into a half cycle each time the thyristor turns off, but the half cycle is ended prematurely when the thyristor goes on again as the c.b. opens.

The operating frequency of the converter thus synchronizes automatically to half that of the c.b. under this condition. Only a fraction of one natural half cycle is needed to charge C_1 to a useful value, so good sparks are generated up to very high engine speeds.

The c.b. bounce-suppressing and pulse shaping section of the unit operates as follows: When the c.b. points are closed, a standing current of about 250mA is passed through

COMPONENTS LIST

Resistors			
In the list below	w the prefix R and the su	ffix I have been om	itted for clarity.
1-50°	4-470	7-2701	10-220
268k	5-3.3M	8-2701	11-100
3—1k	6-1	9-220	12-100
*5-watt wire-wo	und		
t2-watt			
remainder all O.	5 watt		
Capacitors			
C,a and C,b-0.	5µF. 600V working, paper	or Mylar.	

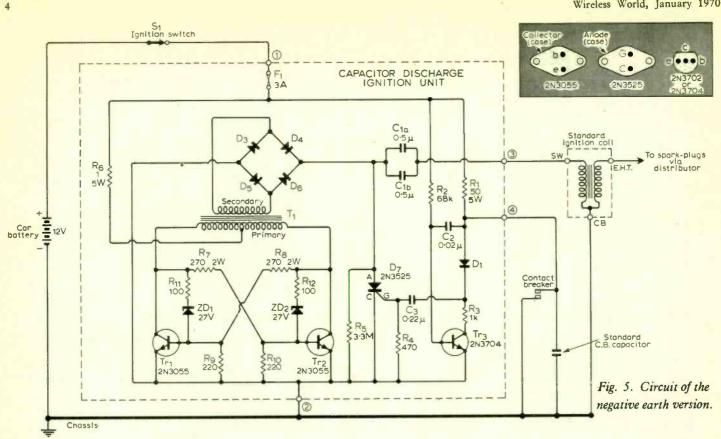
 C_2 —0.02 (F. 50V working, Mylar, C_3 —0.22µF, 50V working, Mylar.

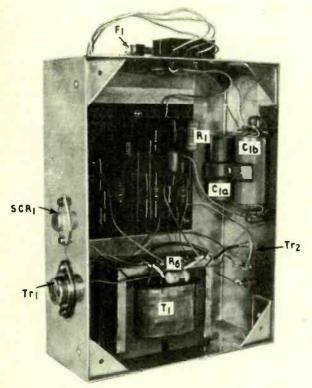
uctors		
	positive earth	negative earth
	2N3055	2N3055
	2N3055	2N3055
_	2N3702	2N3704
	1N4001	1N4001
	1N4001	
	1N4005	1N4005
	2N3525	2N3525
	27V. 5%, 400mW zener diodes	6. E
		positive earth - 2N3055 - 2N3055 - 2N3702 - 1N4001 - 1N4001 - 1N4005 - 2N3525

Transformer 7,

Modified I.t. or battery charger transformer rated at least 30VA. A 240V primary, 17V secondary, 2A tranformer is sultable when modified as per text.

Wireless World, January 1970





Two views of the prototype positive earth version.

the points via R_1 ; the R_1 - D_1 - C_2 junction is at ground, and the R_2 - C_2 junction is grounded via Tr_3 base-emitter junction. Assume that C_2 and C_3 are fully discharged.

At the instant that the c.b. points open, 12 V appear across the points, and C_3 charges rapidly via R_1 - D_1 and the thyristor gate which turns on. Simultaneously, C_2 charges rapidly via R_1 and Tr_3 base so that Tr_3 turns on.

At the instant that the points close again, the R_1 - D_1 - C_2 junction once more drops to ground volts; C_3 is still fully charged, however, and remains so, since D_1 is reverse biased under this condition; C_2 is also fully charged, but, since its R_1 - D_1 side has been pulled down to ground volts, it drives

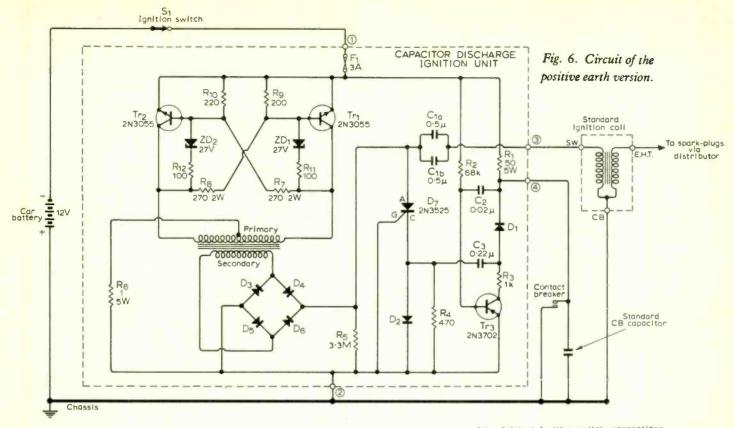
 Tr_3 base sharply negative, so Tr_3 is cut off. C_3 thus has no discharge path at this stage, and retains full charge. Consequently, should the points bounce open again at this stage (point bounce only occurs within the first two or three hundred micro-seconds of initial point closure), the thyristor will not be triggered back on again. Now, as soon as the points close, the C_2 charge starts to leak away via R_2 , and eventually, after about 600μ s, the charge falls to near-zero and Tr_3 is biased on via R_2 . Once it is turned on, Tr_3 provides a discharge path for C_3 via its collector and R_3 and R_4 ; C_3 then discharges rapidly, with a time constant of about 35μ s. At the end of this period, C_2 and C_3 are once more fully discharged, and the thyristor is ready to be triggered on again.

Thus, the thyristor is triggered on as soon as the points open, but can not be operated again until the points open again after being fully closed for at least 600 us. The thyristor is thus immune to false triggering by c.b. point bounce.

The positive ground version of the ignition system is shown in Fig. 6. This is similar to that described above, except that a few circuit polarities are changed and the thyristor is triggered on with a negative pulse applied to its cathode via D_{2} .

The only problem involved in the construction of the unit is that of finding transformer T_1 . This is an iron-cored unit with a turns ratio of 15:1 at a power rating of 30VA or greater, and with a centre-tapped low-voltage winding. The easiest way to obtain this unit is to re-wind an existing l.t. or battery-charger transformer. The winding procedure is very simple, and the following is an account of that used on the prototype:-

The transformer is required, before modification, to have a basic turns ratio of 15:1 or less. Any l.t. or batterycharger transformer that meets this and the 30VA power requirement can thus be used. The prototype unit started life as a 240 V:17 V, 2 A battery-charger transformer, and thus satisfied the above specification. Once selected, the lowvoltage winding of the unit must be re-wound and centretapped to give an exact 15:1 ratio, i.e., a ratio of 240 V:16 V



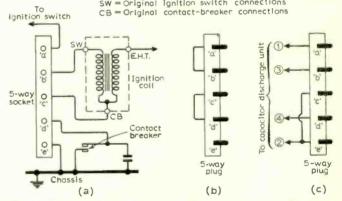
in this particular case.

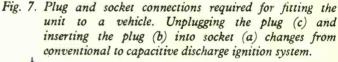
To rewind the transformer, remove its securing clamp and dismantle its iron core laminations (making note of their method of assembly), and then remove the coil bobbin. Next, unwind the entire low-voltage winding (which is invariably the outer winding on the bobbin), and carefully note the total number of turns used; now divide the number of turns by the original value of l.t. voltage, to give the transformers basic turns-per-volt value. On the prototype, total turns were 134, and the original voltage was 17, giving a turns-per-volt value of 7.9. Now calculate the l.t. voltage needed to satisfy the 15:1 final turns ratio of the transformer (16 V in this case) and multiply by the turnsper-volt value (7.9) to give the total number of turns to be rewound (128); now re-wind this number of turns on the bobbin to form the primary of the ignition unit transformer, taking care to make a tap at the half way mark. Finally, re-assemble the core laminations and re-fit the transformer clamp; the transformer is then complete and ready for use. The original mains primary is now of course the secondary of the new transformer.

Construction of the rest of the unit should present no problems, and it can be wired-up direct from the circuit diagram. The prototype positive ground version of the unit (see photographs) is mounted in an $8 \times 6 \times 2\frac{1}{2}$ in metal box; the two power transistors $(Tr_1 \text{ and } Tr_2)$ and the thyristor are mounted, via insulating washers, to the box surface (which acts as a heat sink); most of the remaining components are mounted on a piece of Veroboard Panel; external connections to the unit are made via a 4-way terminal block.

When construction is complete, give the unit a simple functional check by connecting terminal ⁽²⁾ to chassis and terminal ^O to the 'hot' side of the car's battery; a "humming" noise should now come from the unit, indicating that the converter section is operating, and total current consumption should be roughly 800mA; approximately 400 V should be available between the anode and cathode of the thyristor when tested with a 20,000 o/volt meter. If this test is satisfactory, the unit can now be fitted to the car.







The complete unit can be either mounted in the glove compartment (as in the case of the prototype), or can be fixed to the rear fire-wall of the engine compartment (but not close to the exhaust system). The unit can be either wired directly to the existing coil and c.b. assembly, or, preferably, can be wired to these components via a 5-way plug and sockets (Fig. 7), in which case the driver can change from conventional to capacitor discharge ignition by simply fitting an alternative plug into the socket.

Once wiring is complete, turn on the ignition, operate the starter, and check that the system functions well under actual driving conditions; there is no need to re-adjust c.b. or spark-plug gaps, etc.

Results vary from one car to another, but improvements are particularly evident in cars that have covered a considerable mileage since their last tune-up.

Finally, once the unit has been found to perform satisfactorily over a reasonable mileage, it is recommended that the entire circuit be covered with an electrically insulating coat of water-proof paint or varnish, to exclude the harmful effects of moisture. The unit can then be expected to operate correctly for the life of the vehicle.

The Future of Linear I.C.s

A few simple integrated circuits, with guaranteed long-term availability, could meet nearly all the needs of the industrial manufacturer

by R. Hirst*

6

Linear integrated circuit packages have been available for a relatively short period in Britain. By virtue of the techniques employed and the expense incurred in developing and manufacturing monolithic circuitry reasonably large production quantities are required to make the selling price compatible with circuits manufactured from discrete components. This fact in itself restricts the market to which the initial product may be tendered as the industrial design engineer, with a relatively small piece-part requirement, is unable to incorporate devices that are essentially made for the domestic sphere. While the majority of linear integrated circuits will meet industrial requirements, it is the fear of an abrupt cessation in supply at the end of two or three years, due to the biennial change in the requirements of the mass radio and television market, that causes the main concern to the long-term industrial user.

Linear integrated circuit manufacturers seem unable to grasp this situation and continue to pour into the market complex, incompatible and noninterchangeable units mounted in a multitude of mechanical assemblies as can be seen from Table 1.

Based upon a simple survey it would seem reasonable to maintain production of one or two devices that at the present time have more than paid for their tooling costs by virtue of large-scale distribution. It would be necessary to inform the industrial equipment manufacturer which devices would be available for a relatively long time. The consumption, based upon this type of selling, could be surprisingly large and the risk to semiconductor manufacturers spread over a much greater number of customers.

To show how simple linear integrated circuitry could be, the following excursionary appraisal of the requirements of a substantial portion of the industrial consumer has been presented. There would appear to be four main areas in which integrated circuitry could be used to great advantage, these are: switching, 1.f. amplification, h.f. amplification, frequency conversion.

Switching: This appears to have been adequately covered by the majority of

TABLE 1 Characteristics of some i.c. amplifiers					
type	response at – 3dB	gain (dB)	supply volts	pack	
SN777	d.c 100kHz	70	4.5	*	
SN7510	d.c 40MHz	42	+8& -8	*	
SN7510L	d.c 40MHz	42	+8& -8	TO-99	
TAA111	d.c 150kHz	62	7	TO-76	
TAA121	d.c 150kHz	74	7	TO-76	
TAA131	d.c 20kHz	56	5	*	
TAA141	d.c. — 20kHz	56	5	TO-76	
TAA300	d.c 25kHz	50	5 5 9 7	TO-74	
TAA310	d.c 15kHz	100		TO-74	
TAA293	d.c 600kHz	80	6	TO-74	
TAA263	d.c 600kHz	77	8	TO-72	
TAA350	d.c 12MHz	80	6	TO-74	
TAA231	d.c 30MHz	20	12	TO-78	
F104A	d.c 35MHz	20	20	TO-74	
F104B	d.c 45MHz	20	20	TO-74	
CA3011	0.1 - 20MHz	60	10	TO-74	
CA3020	d.c. — 6MHz	58	9	TO-74	
CA3021	d.c 2.4MHz	56	+188-6	TO-74	
CA3023	d.c 16MHz	53	+188 -6	TO-74	

semiconductor manufacturers and it is possible to obtain devices from different manufacturers that are directly interchangeable. The presentation has stabilized in the shape of fourteen-lead dual-in-line packages, usually epoxy encapsulated. The cost of the pack approaches or improves upon the cost that may be achieved by discrete techniques and the only aspect that now remains is to have more standardization and interchangeability.

L.F. amplification: As the majority of integrated circuits are d.c.-coupled, it would seem reasonable to lump together the l.f. and h.f. requirements thus reducing the consideration of linear amplification to a single unit. This device is described under the heading of h.f. amplification.

H.F. amplification: There is an integrated circuit available on the market with a flat frequency response up to 45MHz which is entirely d.c.-coupled. This device is being used in large quantities in the manufacture of domestic radio and television. It is a Mullard unit type F104B and is mounted in a 12-pin TO-5 can. The internal circuit of this device is shown at the left-hand-side of Fig. 1 and it can be seen to be very simple in design but nevertheless adequate in performance. A variety of response curves may be obtained by changing the value of

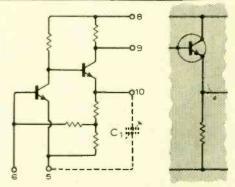


Fig. 1. The Mullard F104A/B is shown on the left which can fulfil nearly all the industrial manufacturers needs for l.f. and h.f. amplification. The shaded area contains an emitterfollower circuit which can be added so that the circuit can be made to drive long coaxial lines.

 C_i as shown in the graph of Fig. 2. The circuit in the shaded portion of Fig. 1 is a directly coupled emitter-follower which can be used to reduce the output impedance so that the unit may be terminated in a coaxial lead to feed a further unit which could be some distance from the amplifier.

This circuit has been used in a number of assemblies operating from 100Hz to 30MHz and it was found that by substantially increasing the value of C_1 that the gain at 100kHz could be

increased from 20dB to 30dB if some increase in harmonic distortion could be tolerated. From the curves in Fig. 2 it is obvious that, if the amplifier is used in a relatively narrow band, the value of C_1 may be altered to accurately provide a given level of gain. For instance with C_1 in the order of 25pF the gain at 50MHz would be approximately 26dB. However, if the value of C_1 was reduced to 12pF the gain would have decreased to something in the order of 19dB. As C_1 is an external component it is an easy task to use a ceramic trimmer adjustable from 5pF to 25pF to take up the gain spreads of the integrated circuit.

Obviously this package does not deliver a great deal of power but it may be terminated in one or more stages to give the required output level. It may also be preceded by an emitter-follower in order to increase the input impedance should the need arise. At frequencies below 10MHz the small value of C_1 is unlikely to promote sufficient change in gain to enable the variable capacitor to take up the spreads from circuit to circuit and it is probable that if a considerable gain change is required at the lower frequencies it will be necessary to alter the value of the series input resistor. However, as previously indicated, if C_1 is replaced by a fixed large value of capacitor the gain can be adjusted over a considerable range, but it then becomes necessary to change a capacitor physically, rather than make a simple adjustment. This amplifier has now been produced by Newmarket Transistors Ltd, under the title MC 809 and is a thick film device mounted in a dualin-line package.

Frequency conversion: Fig. 3 shows a simple ring modulator circuit using four diodes and two transformers. This unit may be manufactured from discrete components but the degree of balance required for industrial applications, over a wide temperature range, cannot be easily achieved unless the diodes are carefully matched and mounted in a common heat-simulating device. With the aid of standard monolithic techniques this type of modulator may be readily presented on one chip thus ensuring that the elements have a similar temperature coefficient and are mounted in close proximity.

At the present time transistor monolithic ring modulators are available on the market as standard units but unfortunately the frequency range is very limited and cannot be considered for highfrequency work.

The circuit of Fig. 3 is not the only method of obtaining balanced frequency conversion but it is a simple device that can have a very wide and flat frequency response providing that the transformers are designed correctly. A unit of this nature has a considerable field of application throughout the military and industrial manufacturing industry.

Conclusion: Little has been done to establish the needs of industrial manufacturers as far as integrated circuits are

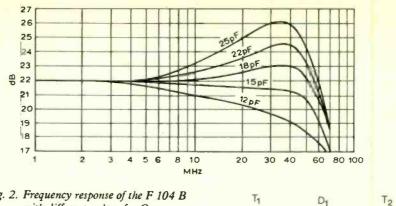


Fig. 2. Frequency response of the F 104 B with different values for C ...

Fig. 3. Basic ring modulator circuit.

concerned. If just one or two of the more simple integrated circuits, at present available to the domestic consumer, were to be classified as devices with long-term availability, the military and industrial manufacturer would undoubtedly respond by including such units in future designs.

During some recent observations into designs promoted by just one industrial manufacturing company it was noted that during the past three years, sixteen totally different amplifiers had been designed to achieve a small signal gain of between 20 and 30dB at 100kHz. Each one of these amplifiers could easily have been replaced by the circuit indicated in Fig. 2 without any detriment to the performance. Some startling but accurate conclusions were reached when cost estimates were prepared for the development and manufacturing cost of the discrete assembly on one hand, and the integrated assembly, on the other.

In the instance of the sixteen amplifiers designed around discrete components, it was ascertained that a total of 2 manvears were involved in the engineering, drawing and planning. The total annual consumption of the final amplifiers was small in the order of 500 units, costing approximately £5 each. The total cost over a manufacturing period of five years was as follows: development-£8,000; 2,500 units-£12,500; giving a total expenditure of £20,500.

Replacing these sixteen different designs with a common integrated circuit the development period could be reduced to one half man-year costing about £2,000 leaving a total of £18,500 to be spread over 2,500 integrated units giving a unit price of £7 8s for a simple threestage device. There is no reason why a semiconductor manufacturer could not make a substantial profit at such an elevated unit price. The advantage to be gained by the industrial manufacturer at such a price would not be directly financial but a reflection in the substantial

period of time that a skilled engineer would now have to devote to the more elaborate task of system design.

Input

f2

The modern circuit engineer is now likely to find that his services are more and more in demand in the laboratory of the semiconductor manufacturer where, chemistry, physics and electronics come close together. There is a vast shortage of skilled engineers in every country and it has to come that the majority of skills available will be employed in the design of systems using integrated circuitry as the basic building blocks. This does not detract from the skill of the circuit engineer but to the contrary indicates that a much higher degree of skill must be used in planning the minimum number of configurations to be used over a very wide and varied market.

* Communication Division, S.T.C. Ltd.

Corrections

J. Dinsdale, author of "A Design in Retrospect" in the November issue, writes: "There is an unfortunate ambiguity in Fig. 5 which does not make it clear whether the mk I or mk II design is being discussed. It is important that the earphone-loading network (shown in the dotted box) is connected in place of the loudspeaker in whichever design is being used. On no account should there be a direct d.c. path from the collector of Tr₆ to ground, as Fig. 5 could imply." Also for May read April in ref. 9.

The values of two resistors in the Wien Bridge Oscillator on page 575, December issue, were incorrect in the diagram. For 68k and 33k, read 6.8k and 3.3k.

See also page 11 for addendum to last month's "Letters"

News of the Month



American radio and TV production

8

It would appear from the latest figures issued by the U.S. Electronic Industries Association that there is a grave decline in the indigenous radio industry. Not that the sales of domestic receivers have declined but that there has been a growing influx of imported sets while the number of home-produced models has decreased.

The total sales of domestic a.m. and f.m. receivers produced in the U.S. during the first nine months of 1969 was 3.58M compared with 4.15M during the same period in 1968. Imports, however, rose by about 5M to 24.6M of which some 4M eventually bore U.S. company labels. The picture in car radio is very different. Of a total of 8.87M units (a slight increase on the 1968 figure for the same period) just over 1.2M were imported.

Of a total of 9.85M television receivers sold during January-September 1969 (of which over 4.6M were colour) 1.23M imported sets bore U.S. labels and a further 1.58M carried foreign labels. Incidentally, about 13% of the colour receivers sold in the U.S.A. during the first nine months of 1969 were imported.

Figures for the disposal of receivers in the U.K. (supplied by the British Radio Equipment Manufacturers' Assoc.) do not show imported equipment. Disposals of domestic radio receivers declined from 762,000 for the first nine months of 1968 to 547,000 for the same period in 1969. Car radio sets dropped from 309,000 to 262,000. Monochrome television receiver deliveries declined from 1,220,000 in 1968 to 1,172,000 in 1969 and colour sets from 89,000 to 77,000.

German satellite earth station

Germany's earth station at Raisting, near Munich, now has a second paraboloid aerial. This, like its counterpart at the U.K. Goonhilly station, will enable communication to be maintained via satellites in both the eastern and western hemispheres.

The main physical difference between Raisting's aerials I and II is that the designers have dispensed with the use of Raisting's second paraboloid. In the background is the radome of the first aerial

the radome cover in the latest installation. Although this gave protection from the weather it also created a problem-a film of water or ice on the radome caused background noise. Aerial II is fitted with 5000 infra-red radiators to prevent icing. The dish is 28.5m in diameter and the gain 60dB which corresponds to a power gain of one million as compared with an isotropic radiator operating at 4GHz. Maser pre-amplifiers with a 25-MHz bandwidth were originally used in the receiving section of the earth station, but parametric amplifiers with a 500-MHz bandwidth and a gain of 10,000 have now been installed by Siemens who undertook the refurbishing of the station.

Information service for engineers

INSPEC the Institution of Electrical Engineer's information service in physics, electrotechnology and control, is to launch a selective dissemination of information (SDI) service in electronics in January. It will be available on an individual or group subscription basis in the United Kingdom only. Periodical articles on all aspects of electronics, published in English or English translation, will form the basis of the service. The institution plans to start a comprehensive SDI service covering all languages and the complete subject range of INSPEC in 1971. This service is part of the overall plan for the development of comprehensive information service, which is being supported by the Office for Scientific & Technical Information of the Department of Education & Science. For the past year the SDI service in electronics has been limited to some 600

research workers as part of a governmentsupported information research project. Further information and details may be obtained from the manager, INSPEC SDI Investigation, I.E.E. 26 Park Place, Stevenage, Herts.

Cranfield Institute of Technology

Cranfield College of Aeronautics, which was founded at Cranfield, Bedford, in 1947, has been granted a Royal Charter to become the Cranfield Institute of Technology with power to award its own higher degrees. As its original title implies it has been concerned principally with aeronautics but in future its object will be "to advance, disseminate and apply learning and knowledge in the disciplines of the sciences, engineering, technology and management". The Institute will also pay particular attention to "the educational needs of industry, commerce and the public services".

Laser space communication

The first laser communications system to be used in a satellite is to be developed by Aerojet-General Corp., of Azusa, Calif., under contract to NASA. The equipment is to be used aboard the Applications Technology Satellite—F (ATS) which is scheduled for launching from Cape Kennedy into a synchronous orbit in 1972. The contractors will develop both the spacecraft equipment and the associated ground equipment. When ATS—G is launched in 1974 the laser communications experiment may be extended to include spacecraftto-spacecraft links.

Conferences on tape

"Cassette Colloquia" is the name of a programme begun by the Institute of Electrical and Electronics Engineers to keep its members technically up to date. Cassette recordings of special seminars, workshops, sessions etc, conducted by the I.E.E.E. will be available to members and non-members. The recording technique for the cassettes involves speech compression without pitch change. This, in conjunction with editing, allows $2\frac{1}{2}$ hours of material to be converted to 75minutes in the cassette of a recent meeting. A cassette containing this length of recording costs \$10.

V.H.F. complaints

The B.B.C. has completed an analysis of reports of unsatisfactory v.h.f. reception (during 1968/69) which shows that more than 50% of the complaints were due to the use of inadequate aerials or to faulty or maladjusted receivers. A great deal of dissatisfaction could be avoided, it is said, if dealers would advise when an external aerial is necessary and would also make sure that customers know how to tune their receivers.

Full colour spectrum from infra-red

New phosphors, employing rare earth elements in crystals, have been found by workers at Bell Telephone Laboratories to convert infra-red radiation into any colour of the rainbow. The source of infra-red energy is a gallium arsenide diode. The initial use of the combination of GaAs diodes and infra-red-to-visible phosphors was reported by General Electric. The phosphors can be painted on the diodes-green or red light is produced by certain crystals containing erbium or holmium, and blue light using thulium. With one of the phosphors, colours gradually change from green through yellow, off-white, and orange, and finally to red, as power is increased. The red light so produced is as bright as that emitted directly by other solid-state lamps.

WWV standard frequency transmissions

The National Bureau of Standards (U.S. Department of Commerce) in Boulder, Colo., is responsible for the operation of four radio stations (WWV, WWVB and WWVL at Fort Collins, Colo., and WWVH, Hawaii) that transmit accurate time and frequency information. The formats of two of these stations, WWV and WWVH, are being reviewed for possible changes and modification. A questionnaire has been sent to many known users of the broadcast services and any other users (government, military, industrial, scientific or private individuals) who wish to receive the questionnaire are asked to write to WWV 1969, National Bureau of Standards, Boulder, Colo. 80302.

For the record, station WWV has been transmitting standard radio frequencies on a regularly announced schedule since March 1923 and WWVH began supplementing the broadcast services of WWV from a site on Maui, Hawaii, in 1948. Both stations broadcast the same services on high-frequency carrier waves. Stations WWVB and WWVL at Fort Collins transmit on l.f. and v.l.f. The services of all four stations are described in publication 236, NBS Frequency and Time Broadcast Services, available from the Superintendent of Documents, U.S. Government Printing Office, Washington, D.C. 20402, for 25 cents.

TV camera for low-light levels

A television camera, type GTNV-1 capable of producing pictures from scenes illuminated at light levels equivalent to starlight, has been introduced by STC. Minimum sceneillumination requirement for the camera, is about 2×10^{-5} ft. candles so that it can respond to scenes that are invisible to the human eye. A vidicon tube is used in conjunction with a three-stage image intensifier having a very high overall gain. Typical brightness magnification is 35,000 times. No especially contrived illumination such as infra-red beams, or reliance on self-emitted infra-red is necessary. For use underwater, where there is very little light, a clearer picture is obtained by using the natural light available rather than an artificial source, the light from which tends to be scattered back to the camera and so degrading the picture.



The STC low-light television camera with cover removed

W.W. Diary

The larger-page size 1970 Wireless World Diary (5×3 inches) has enabled a more readable type to be used for the information section which includes those features found to be most acceptable to users. They include formulae, circuits, aerial data, colour television and stereo broadcasting characteristics, transistor data, frequency allocations, addresses of organizations and many other facts and figures.

The Diary, which has a weekat-an-opening, costs 10s (leather) or 7s (rexine).

American incentive licensing allocations

Although the U.S. Federal Communications Commission recently suspended the application of proposals for increasing the sub-allocations of h.f. bands available only to amateurs holding Extra Class licences, there remain substantial portions of the bands available only to those holding Extra Class and Advanced Class licences, as part of the policy of encouraging American amateurs to study for the more advanced licence examinations, in a scheme introduced in November 1969. The first 25 kHz of the 3.5, 7, 14 and 21-MHz bands are available only to Extra Class telegraphy; the frequencies 3.8 to 3.9, 7.2 to 7.25, 14.2 to 14.275 and 21.25 to 21.35MHz are now all subject to reservations for either Extra Class only or for Extra Class and Advanced Class telephony. Our correspondent Pat Hawker says British amateurs have expressed opposition to an A.R.R.L. proposal that American telephony operation should be authorized in the band 14.1 to 14.2MHz.

Mysterious generation of u.h.f. exploited

An unexplained phenomenon, discovered at RCA Laboratories in 1967, has been harnessed by RCA to produce the most powerful pulses of radio energy in the u.h.f. range yet achieved by a solid-state device. The effect occurs in avalanche diodes, when they are placed in a circuit tuned to oscillate at frequencies lower than those at which the diodes are supposed to be able to oscillate. For reasons that are still not fully understood, when electrical pulses are now applied to the diodes, they abruptly enter an "anomalous mode" of operation and begin to produce microwave oscillations with powers and efficiencies substantially higher than normal. It is reported that by combining five such devices in a single tiny package and operating them in the anomalous mode, microwave pulses with peak powers above 1,200W have been powers above 1,200W have been produced with efficiencies above 25%.

approximates to-

$$M \approx \frac{h_{fe} \cdot R_L}{h_{ie}}$$

The Editor does not necessarily endorse opinions expressed by his correspondents

Letters to the Editor

Transistor Distortion

Characteristics

Mr. Linsley Hood's results (given in his November article) are rather unexpected, in one respect. A perfect transistor would exhibit a voltage gain independent of h_{fe}, and, in his particular test circuits, independent of the load resistance as well. Yet the reported figures, even allowing for the imperfections of the transistors, are at variance with this expectation. They are also at variance with my own measurements.

In a practical circuit, the voltage gain is very nearly g_m . R_L , for a planar transis-tor with low "extrinsic base resistance". Now g_m is a function of the collector current, not the current amplification factor. It is about 40 Ic. Thus a transistor operating at $I_c = ImA$ has a g_m of about 40mA/V.

The voltage gain is therefore $40.Ic.R_L$. In Mr. Hood's tests, Ic. RL was kept constant, so one would have expected the voltage gain to be constant, to a first approximation, irrespective of the variations in hfe, Ic, and RL. In circuit A, for example, $I_C.R_L$ was fixed at 5, so the expected voltage gain is 200, not 40-140 as reported.

My own quick tests on a silicon planar transistor (BFY51) produced the expected results: the voltage gain with 5V dropped across the load was 185-210 for loads of 1-8.2kn; i.e., and 8 to 1 variation in collector current. There is clearly something wrong somewhere.

In the case of alloy transistors, the extrinsic base resistance is comparatively high, and changes the performance appreciably. At the higher collector currents and lower values of hfe, this resistance (perhaps a few hundred ohms) is comparable with the "true" input resistance 25. hfe/IE. Its effect is to reduce the apparent g_m and also to make the transistor operate, not as a purely voltage-driven stage, but in a mode between voltage drive and current drive. This latter effect improves the linearity It follows that the linearity of any voltage amplifier stage can be improved by inserting base resistance. The price you pay is in reduced gain and increased noise. (Much the same effects are obtained by the use of an unbypassed emitter resistance.)

If large output voltage swings are taken, distortion due to Early Effect may become important. (This was reported by Dr. Bailey in connection with one of his power amplifiers, where the driver stage had to deliver large swings.) It may well be that a low-hfe transistor shows less of this distortion than a highhe one, though correct selection of types is perhaps better than selecting for low hje.

G. W. SHORT, South Croydon, Surrey.

The author replies

I was pleased to read Mr. Short's letter, and I note with interest, his argument that a transistor should, ideally, always give an identical stage gain, as a voltage amplifier. However, this is not the situation one finds in practice, nor is it the conclusion one draws from gain calculations made using the classical formula,1 using the conventional h parameters, for a common emitter configuration.

$$M = \frac{1}{h_{re} - \frac{h_{ie}}{Z_L} \left(\frac{1 + h_{oe} Z_L}{h_{fe}}\right)}$$

assuming $Z_{gen} = 0$. Taking the transistor type which he quotes, and obtaining the typical values for the h parameters, hie, hoe, hre and hfe from the Mullard data sheets, the calculated stage gains for a BFY51, under ideal conditions of zero source and emitter impedance, vary from 210 to 319 over the range of collector loads I to $IOK \Omega$.

However, there is a less complex formula quoted by Manasse², using the concept of the "h determinant" Δ_h , (Δ_{he} is equal to hie. hoe-hfe. hre),

$$M = \frac{h_{fe} \cdot R_L}{R_L \cdot \Delta_{he} + h_{ie}}$$

Since over the range of loads in question with a BFY51, Δ_{he} is very small, this

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$$1 \approx \frac{h_{fe} \cdot R_L}{h_{ie}}$$

So, if the input impedance of the transistor increases linearly with the product hfe. RL (and hfe may remain nearly constant), the theoretical condition could be met. Normal device shortcomings, such as doping inhomogeneity, carrier trapping and the base-emitter spreading resistance presumably give rise to the failure of the theoretical model.

However, with regard to the gain figures I quoted for the devices I examined, it had not been my intention that these values for gain should be taken as the voltage gain of such devices under ideal voltage amplifier conditions. Alas, low-distortion signal generators do not have zero output impedance. My intention was, rather, to establish a form of "figure of merit" for such devices, and to determine the comparative performance, say, of germanium versus silicon and n-p-n versus p-n-p. In this context the fact that the signal generator had not a zero impedance output was not of importance.

In fact, the apparatus used was a Solartron VF252 precision millivoltmeter, a Radiometer BKF5H distortion meter and a Marconi TF1101 low-distortion oscillator, with 1kHz output filter. (The modulus of the output impedance of the Marconi oscillator is 660 ohms, which accounts for the actual stage gain being lower than the calculated zero input impedance value.) It was remiss of me not to mention in the article the source impedance used, but, surely, if one really wanted to know what the typical stage gain of a particular device would be under zero input and zero emitter resistance conditions, one would calculate it from the formulae, rather than try to measure it with a possibly very untypical component.

With regard to the point raised by Mr. Engstrom in his letter in the December issue, may I say that the points he raises are agreed. The treatment of transistor voltage amplifier non-linearities on the basis of variation in the input admittance is, indeed, the classical approach. However, I quote Mr. P. J. Baxandall's observation that in transistor circuit design it is much more fruitful to consider the devices as voltage amplifiers; and treat their non-linearities on that basis, rather than to endeavour to swamp the input impedance changes by the inclusion of massive input or emitter circuit impedance.

J. L. LINSLEY HOOD

References

- I. "Transistor Circuit Design", Walston and Miller, pp. 98-99.
- 2. "Modern Transistor Electronics, Analysis and Design", Manasse et al, (Prentice Hall), pp. 46-49.

Stereo gramophone pickups

The most interesting and timely article by Mr. Stanley Kelly on stereo gramophone pickups in the December issue prompts me to raise two points. First, although in reviewing the dynamics of the transducer-stylus-groove system, Mr. Kelly does assume a compliance for the disc material of 3×10^{-8} cm/dyne, neither he, nor any other authority that I am aware of, tells us very much about the behaviour of disc material. What, for example is the effect of temperature upon it and are there significant differences between various record manufacturers' products in this respect. I have long felt that the characteristics of record material and, above all, resonances which occur within the disc and the effect that different modes of bedding records on the turntable have upon this, are worthy of close examination.

Secondly, Mr. Kelly's reference to novel principles for pickups (straingauge and photo-electric) recalls another possibility which must be of particular appeal to readers of Wireless World, because its life-force is h.f. The mono version of this type of pickup uses a conventional stylus flexibly anchored at the rear end carrying, instead of coils or magnets, a simple vane of quite negligible mass. The movements of the stylus due to the groove modulation causes relative movement between the vane and a fixed plate or electrode which is continuously energized to emit a constant high frequency e.m.f. The stylus-driven vane is connected to a tiny "receiver"-merely a tuned miniature pot-core inductance, and semiconductor diode. The amount of h.f. energy reaching this receiver at any instant depends upon the instantaneous position-hence impedance -which the vanc forms with the fixed plate or pole; in short the h.f. energy input is amplitude modulated by the movements of the vane. The diode delivers an audio-frequency product to a load resistance of $100k\Omega$, or so, and this is conducted away to the input of the record player amplifier.

The system is readily adapted to stereo by the employment of two vanes at right angles. The same h.f. pole is easily adapted to energize both vanes; the receiver and diode are, of course, duplicated-one for each vane; and the a.f. currents from them are the two inputs to the stereo channels. Unaffected by d.c. or 50-cycle a.c. fields-magnetic or electrostatic-and potentially capable of outputs normally met with in crystal pickups, the advantages are evident and more than overcome the need for the oscillator necessary to energize the "pole". Such an oscillator can easily be housed within the pickup arm if need be. Indeed, why should we not go the whole way to achieve the ideal completely conductorless pickup arm-no filamentary wires, no mercury baths-just two oscillators in the pickup head, frequency-modulated by

the same two moving vanes and one common fixed plate; two transmitters, in fact, transmitting over a few inches to two miniature f.m. receivers strategically placed at the side of the turntable. H. J. N. RIDDLE, Sherborne, Dorset.

The author replies

With reference to Mr. H. J. N. Riddle's letter, the values quoted for disc material were obtained by direct measurement by myself, and are indicative of present day vinyl products. For any given record material the absolute value of compliance is a function not only of the record material, but also of stylus radius, stylus pressure, the mechanical impedance with reference to the stylus tip, and temperature; it not only varies between record manufacturers but to a second order from batch to batch of record material.

Additionally, a given record does tend to harden with time and over a period of years the effective compliance decreases. Although one tacitly assumes that the impedance of the record as seen by the stylus tip is a pure compliance, this is more of a pious hope than fact. There are the inevitable loss components and the variation of impedance with frequency suggests a more complex structure than a simple RC combination.

I agree with Mr. Riddle that the characteristics of the disc material are taken too much for granted and although at various times I have investigated particular facets of this interesting problem, the finances of a private laboratory cannot unfortunately accommodate the detailed investigation that this subject requires.

With reference to the stereo capacitor type of pickup, one such unit has been produced in Japan and during the past few years I have received samples of this device. Unfortunately, the first was damaged in transit, and the second unit exhibited lack of stability, presumably due to frequency variation of the h.f. oscillators. Additionally, the setting up of the two transmitters and receivers was critical if audible beat noises between the two systems were to be eliminated. The mechanical characteristics of the moving vane are not simple, acceptable signal-to-noise ratio controls the minimum change of capacitance and this in turn determines the vane's dimensions, which are larger than would have been expected.

I agree that on paper at least the variable capacitance type of pickup is very attractive, expecially if one can fit the complete transmitter within the pickup head shell. My own thoughts on the matter are to use a single r.f. transmitter phase modulated by each channel with suitable decoding systems. STANLEY KELLY

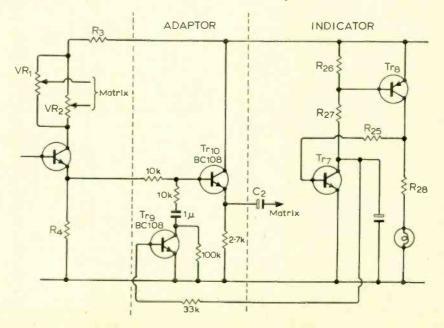
Stereo decoder adaptor

We regret that the circuit was omitted from the letter under the above heading on p. 565 in the December issue. To avoid ambiguity we are reprinting the letter with the diagram.

Having seen the circuit for a stereo decoder adaptor in "Circuit Ideas" in the September issue I am prompted to send you a much simpler circuit which I have been using for several months.

Instead of having a variable gain amplifier on each channel a 0.5 attenuator is placed before the decoding matrix when a mono programme is being received. With a mono signal Tr_9 saturates and earths one end of the attenuator chain formed by the two 10k Ω resistors. Thus only half the mono sign is applied to the matrix. On stereo Tr_9 is off and the full signal is applied to the matrix. The 100k Ω resistor ensures that the 1 μ F capacitor does not have to change its charge when going from mono to stereo.

A. ROYSTON, University of Warwick, Coventry.



Circuit Ideas

Constant current generator

Dual transistors are used to ensure cancellation of thermally variant transistor parameters in the simple constant current generator shown.

By suitably proportioning the resistor values the drift of the constant current due to variation of transistor base-toemitter junction voltage, collector-tobase leakage current, and current gain are cancelled out.

For cancellation of base-to-emitter voltages the necessary circuit relationship is:

$$R_3 = R_4.$$

To assist stable operation with change of leakage current and gain the remaining resistors are proportioned to satisfy the equality:

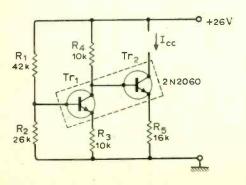
$$R_5=\frac{R_1R_2}{R_1+R_2}.$$

The output current is given, with good accuracy, by:

$$I_{cc} = \frac{V_{cc}}{R_5} \left(1 - \frac{R_2}{R_1 + R_2} \right)$$

The transistors should be operated with equal emitter currents to ensure tracking of the two base-to-emitter voltages with temperature.

The resistors should be wire-wound types with low temperature coefficients. The differential temperature coefficient of the resistor pairs R_1 , R_2 and R_3 , R_4 is more important than the absolute temperature coefficient.



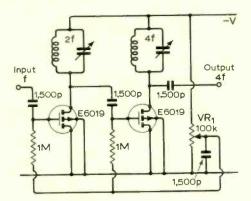
The measured temperature coefficient of I_{cc} is typically 0.0015%/°C over the temperature range 0 to 100°C.

As shown the constant current is directly proportional to the supply voltage. To make the constant current independent of the supply, resistor R_1 may be replaced by a temperaturecompensated zener diode. An additional resistor (R_b) is then required in series with the base of transistor Tr_1 , such that $R_b = R_5$.

M. CADWALLADER, London N.W.3

A M.O.S.T. frequencydoubler chain

The high impedance of m.o.s. transistors allows tuned frequency doublers to be cascaded without requiring impedance transformation. The use of enhancement devices eliminates the need for a separate bias supply. The arrangement is shown in the diagram. The potentiometer is adjusted to provide the required bias



for maximum efficiency. Since C_{iss} is not highly dependent on the applied gate voltage, adjustment of VR, does not detune the preceding circuit. Doublers are not prone to feedback and interaction problems so that dual m.o.s. devices such as the Marconi-Elliott E6029 could be used. The circuit will operate up to 150MHz with the high g_m devices of the E6019, E6029 series.

J. A. ROBERTS, University College, Swansea.

Negative resistance of transistor junction

If the emitter-base junction of a silicon planar transistor is reverse biased, it behaves as a zener diode, with a typical breakdown voltage of 7-10V. If the base is left open-circuit and connection is made to the collector instead, the "zener diode" has a negative resistance characteristic. The effect is exhibited by most

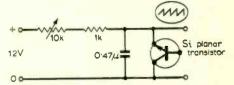
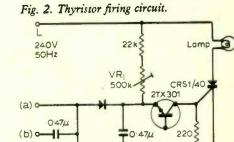


Fig. 1. Test circuit giving ramp waveform.



n-p-n silicon planar transistors, but by few p-n-p types. The relaxation oscillator in Fig. 1 can be used as a test circuit.

The device is also useful for firing thyristors. Fig. 2 shows a simple half-wave lamp-dimming circuit which can be controlled manually, by VR_1 , or by a d.c. input 0 and 10V at (a) or an a.c. input 0 to 10V peal-to-peak at (b). J. A. H. EDWARDS,

Leicester.

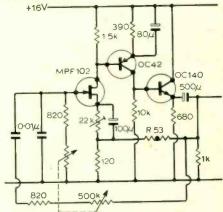
6N

Low-distortion 30Hz-20kHz oscillator

The circuit is a Wien bridge oscillator employing an f.e.t. to reduce damping on the bridge and allow the use of a $500k \Omega$ twin potentiometer. Harmonic distortion of the prototype was reduced to less than 0.05% over the whole band with the aid of the $22k \Omega$ preset resistor. C. A. PYE,



Warwickshire.



Wien bridge oscillator covering audio range with no capacitor switching. (C. A. Pye)

Amorphous Semiconductors

An electronic engineer's view after a recent conference at Cambridge

by J. E. Carroll. Ph.D.

About 350 pure and applied physicists with a sprinkling of electronic engineers were in the Cavendish Laboratory, Cambridge, from September 24th to 27th," to discuss the amorphous and liquid state. Amorphous materials with their lack of obvious structure have for a long time posed fundamental problems of description to the pure physicist, but the applied physicist and engineer are not attracted to a field unless they have a whiff of a practical application. Although negative resistance and switching effects have been reported as long ago as 1962¹, this whiff of practical utility was not scented by the technical hounds until 1968 when Ovshinsky² published a letter entitled 'Reversible electrical phenomena in disordered structures'. The title appears harmless but the contents suggested the use of amorphous semiconductor material in the application of switches with a high ratio of 'on' to 'off' impedance and also in the application of memory devices. Applications to the communications industry and logic functions in computers are then obvious if the device is a success. At first sight there appear to be several useful technological features: apparent lack of sensitivity of the material to small amounts of impurity, the devices can be used in thin film form that would be compatible with modern integrated circuit technology, no power consumption to maintain the memory, to name a few of the more obvious advantages. This then accounts for a small technical explosion of interest in this field. This article attempts to give a simple account of these amorphous semiconductors and their associated potential devices in the light of the recent Cambridge conference.

The amorphous state

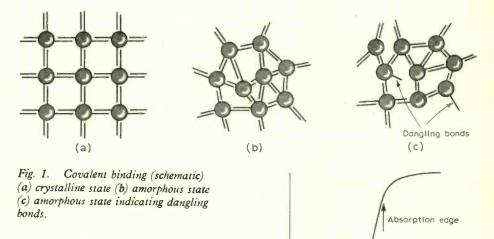
First, let us ask the question, what is an amorphous material? An initial definition would be a material that exhibited no structure or order. So no matter where we looked, we should see a random spacing of atoms in the material.

International Conference on Amorphous and Liquid Semiconductors, sponsored by the Royal Society. The proceedings to be published in a special volume of the *Journal of Non-crystalline Solids*, published by North-Holland Publishing Company, Amsterdam.

Although partially true, this is too naive. The solid material has to be bound together by some cohesive force. This cohesive force can then impose constraints on the extent of the disorder. It is well known that the outermost electrons (valence electrons) of any atom determine the chemical properties of the atom, or in other words determine how one atom binds itself to any other. One type of binding together of atoms is called covalent binding. The valence electrons are shared between pairs of neighbouring atoms, lowering the potential energy of the pair and so binding them together. This binding can extend throughout the crystal. Fig. 1(a) indicates such a scheme for say crystalline germanium. Each atom shares four electrons with neighbouring atoms and this completes a relatively stable configuration. In Fig. 1 (b) the same scheme is shown in a disordered array but to preserve the binding the bonds are still linked. The technical jargon says that the co-ordination number of each atom is preserved. This imposes constraints on the short-range order (say over a couple or so of atoms spacing). To appreciate

this fully one needs to go to quantum theory of electron orbitals, but this is not necessary here. However, over larger distances disorder prevails and the atom spacing and positioning become quite random. A slightly more realistic model is obtained by allowing for several of the valence bonds to be broken, or dangling as they are often referred to. This is shown schematically in Fig. 1(c). It is such covalent amorphous materials, but with more complex structures, that have been causing most interest since in many ways they behave like intrinsic conventional semiconductors. There are, of course, other amorphous materials such as amorphous metals. In this latter case the atoms are bound together by a sea or jellium of almost free electrons shared between a large number of atoms. The binding imposes no long- or shortrange order and the conduction is not significantly changed between the crystaline and amorphous states. At present these latter materials are not of interest.

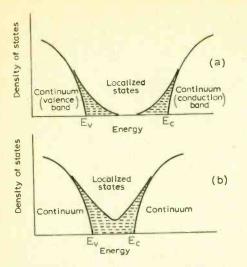
Returning to the covalent amorphous semiconductors, we find further evidence for the short-range order in the absorption of certain wavelengths of light. If a



Absorption

Fig. 2. Absorption edge (schematic). Absorption of light by a crystal against light frequency.

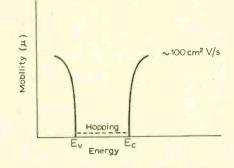
Wireless World, January 1970



14

photon of light has sufficient energy to break a valence electron bond in a semiconductor then it can become strongly absorbed by the material. Thus if the absorption of light shows a marked edge to it as the wavelength of light is changed (Fig. 2) then this is evidence of a uniform binding energy throughout the body of the crystal. Qualitatively similar optical effects are found in amorphous and crystalline material, though often with different magnitudes. From the existence of these effects it can be inferred that there is considerable uniformity of the electronic structure close to each atom. In other words there is indeed a shortrange order in the amorphous material. Now as one of the speakers at the conference asked 'what is a forbidden gap but the binding energy of the valence electrons?' The absorption edge energy is indeed one of the ways of measuring the gap between the valence and conduction band energies in conventional semiconductors. Thus we still expect to find a similar gap in amorphous semiconductors. However, although all gaps are forbidden, some are more forbidden than others! In amorphous material there are lots of broken valence bonds. Although annealing the material can reduce their number, their density is still very high. It is so high in fact that any 'free' electrons, introduced by impurities (on the classical semiconductor basis of creating electrons) become trapped in these dangling bonds. As a consequence the conductivity of the material is not altered even by an appreciable amount of impurity in the material. This is in complete contrast to conventional semiconductors. These traps have a continuum of energies and can in some cases fill up the conventional energy gap. Thus the density of electron states available to electrons with different energies can be drawn schematically as in Fig. 3(a). There is a continuum of states as found in conventional semiconductors but also a high density of localized states forming tails to the continuum. These localized states are in effect the broken valence bonds discussed above. To the electrical engineer who uses frequency filters these tails (or tales) appear most plausible! If one randomly alters the inductances and capacitances of the periodic chain (Left). Fig. 3. Density of electron states in amorphous semiconductor (a) nonoverlapping localized states (b) overlapping

localized states. (Below). Fig. 4. Electron mobility against energy.



of these elements that form a frequency filter, then one finds that the cut-off frequencies become diffuse and propagation is possible for regions extending into the formal stop band of the filter. Thus changing the periodic structure of the crystal in a random way would be expected to allow a certain amount of propagation of the quantum electron waves outside their normal permitted range of frequencies or energies. The farther away from the formal permitted energies the more likely are the states to be localized (in the filter analogy the states are a result of local resonances in the filter structure). These tails of localized states extend both from the band of valence electron energies and the conduction band energies. In some materials these tails can overlap (Fig. 3(b)) so that the whole gap is filled with traps. This latter picture is believed to be the relevant one for the glasses that exhibit switching and memory³.

Conduction can occur as in a con-ventional semiconductor. The electron moves in an electric field and gains energy, thus moving to a higher energy state. This is readily possible for electrons in the conduction band where there is a continuum of empty states above the electrons' particular state. It is also possible in the valence band provided that there are vacancies, or holes as they are called, in the occupation of the upper electron states in that band. These holes then permit the electrons to gain energy in an electric field and so allow conduction. An important difference between the amorphous and crystalline state is the magnitude of the mobility. The increase in disorder implies that the electrons have many more collisions as they move. For a given field the electrons' drift velocity is a good order of magnitude lower than in the useful crystalline semiconductors. The mobility is then around 100 cmV/s² for these amorphous materials. There is a second mechanism that is called 'hopping'. In the high density of localized electron states in the gap, electrons can hop from one state to another under the action of an electric field. However, this hopping process results in a negligible mobility. We therefore arrive at the picture of a 'mobility gap'. Although there may be a high

density of electron states throughout the energy 'gap' in amorphous materials, none the less the mobility of any electrons filling those states can probably be ignored. The idea is shown schematically in Fig. 4.

Since any free electrons become trapped by the unfilled valence bonds one is not surprised to find that these semiconductors only exhibit what is termed an intrinsic conduction. Conduction only occurs in proportion to the amount that the thermal agitation can free electrons from their bonds. The addition of impurities makes little effect on this process. The classical behaviour of an intrinsic semiconductor's conductivity is given by $\sigma = \sigma_0 \exp(-(\Delta E/2kT))$ where ΔE is the band gap energy, or close to this value. This result is also found for amorphous semiconductors of the covalent type. An important consequence of this intrinsic behaviour is that the resistance of any specimen decreases as the temperature increases (more electrons produced to conduct electricity by more bonds breaking). This leads to thermal runaway under some conditions and in turn can lead to negative resistance and switching.

Practical devices

Let us now describe two types of device that were being demonstrated in experimental form by Energy Conversion Devices at the Cambridge conference. They both use films (circa 1 micrometre thick) of an amorphous semiconductor known as a chalcogenide glass. The first type of device is the Ovonic Threshold Switch, or O.T.S., named after Ovshinsky who first reported it in 1968. The threshold voltage is around ten volts and the off resistance can be as high as tens of megohms. As the threshold voltage is reached so the current rises to a few microamps and then switches to many milliamps. The switching time can be extraordinarily fast and this leads to problems in surge currents through the device that can degrade the performance if they are not limited. In the 'on' state the device impedance drops to around 100 ohms (all these figures depend on geometry and so must only be taken as indicating orders of magnitude). The device then remains in the on state provided that the current is above a minimum sustaining value of around 10 mA, or equally the device voltage does not fall below about a volt. If the current does fall below this sustaining level then the device reverts to its high resistance state. Provided that surge currents are limited it is claimed that these switches can be recycled almost indefinitely. The characteristics for the O.T.S. are shown schematically in Fig. 5(a). It should be pointed out that there is evidence against the existence of a closely defined threshold voltage since some workers find that this varies statistically from one switching operation to another⁶. This point was hardly made at the Conference.

Closely allied to the O.T.S. is the O.M.S. or Ovonic Memory Switch. Energy Conversion Devices were exhibit-

Wireless World, January 1970

ing an experimental thin film array of these devices. In this type of device the conducting state of the glass is permanently, although reversibly, changed by the application of the switching voltage, which must be maintained for a time measured in milliseconds. The switch will then move to its low resistance state and remain in this state even though the current and voltage are removed. The device can then form part of a memory store. The low resistance state may be changed back by applying a current of around an ampere for about 100 microseconds. This then restores the device to its high impedance state. Fig. 5(b) indicates schematically the action of this device. To read the state of the memory one applies a voltage from a source with a medium impedance. In the low impedance state the fraction of the voltage dropped across the switch is negligible and so a voltage sensor across the switch can register zero. In the off state the full voltage appears across the switch so the voltage sensor indicates a unity value. This 'read' process can be made extremely rapid so that at present the memory could find applications in a 'read mostly' or 'read only' type of memory store. Ovshinsky used a material with a composition of Te₄₈As₃₀Si₁₂Ge₁₀ for his switches and reduced the arsenic content to around 5% for the memory devices. However. it is not known or, at least, reported what determines whether a device will be a memory, a threshold switch, both or neither!

The chalcogenide glasses used for these devices are covalent amorphous semiconductors and so exhibit a negative temperature coefficient for their resistivity. It is natural to think of the switching as possibly being caused by thermal runaway. Indeed at the Cambridge conference evidence was presented that showed the thermal runaway model fitted several experimental facts. It may, at first sight, be thought that such a mechanism could not account for switching in the subnanosecond speeds that are observed with these devices. However, although the speed of switching is fast, there is a delay of the order of a microsecond before the actual switching occurs. Moreover, it is known that a device with a negative temperature coefficient of resistivity will form a current-controlled negative resistance and in these types of negative resistances the current tends to flow in filaments. This bit of physics can be qualitatively understood by considering a set of parallel and equal negative value resistors. If one resistor takes slightly more than its fair share of current then its resistance falls and it will take more current and so on until all the current is going into the one resistor. Filament formation can imply that the heat required for increasing the conductivity in the filament need only be very small. However, although the thermal runaway theory fits many facts, Professor H. Fritzshe maintained that even with filament formation there was not enough heat for the filament to reach the

required temperature to explain its low resistance, as measured experimentally. It may be that heat causes some slight reversible structural change so that the conduction is no longer intrinsic. Professor H. K. Henisch in another paper suggested that electrical charge effects of the carriers could account for the switching with the current maintaining a plasma in the switch when in the on state. The neutral plasma of charge carriers could imply a high current but low voltage while recombination of the holes and electrons would imply that a minimum sustaining current was required for the plasma. Elsewhere Professor Sir Nevill Mott^{4,5} has suggested that tunnelling of charge carriers through Schottky barriers set up at the electrodes could result in switching. It is safe to say that at present there is no definitive theory on the switching effects in these glasses and indeed it may be a combination of effects is required to explain the facts.

The memory type of device is almost certainly connected with a structural change of the amorphous material caused by heating in a filament. A very beautiful bit of evidence for this theory was given by Dr. C. Sie of Energy Conversion Devices in a film shown to the conference delegates. In a particular material $(As_{35}Te_{35}Ge_{10})$ the switching time is very slow and Dr. Sie filmed the device under a microscope and showed the filament growing from the anode. On first applying the voltage, in excess of the switching threshold, the surface of the semiconductor changed its reflectivity slightly accompanied by a current rise. But then one saw (Fig. 6) a filament growing slowly from the anode towards the cathode contact of the device. As the filament moved towards the cathode the threshold voltage for current switching decreased until, when the filament had fully formed, the threshold voltage was zero and the device was perfectly ohmic in a low resistance state. A microprobe analysis of the composition of the filament showed that it had changed its composition from As55 Te35 Ge₁₀ to As₃₈Te₅₈Ge₄. Temperature analysis with a micro-radiometer showed that the material heated as the current initially started but that as the filament passed under the radiometer the temperature dropped. The velocity of propagation of these filaments could vary depending on material. Rough orders of magnitude suggested the variation was between 100 cm/sec to 10⁻² cm/sec.

Conclusions

It is clear that much technical, technological, and theoretical work remains to be done with many elegant experiments along the way. Some elementary ideas are clear for the amorphous semiconductors but the rigorous theory on which to base quantitative work is lacking. For the practical devices the mechanisms by which they work are only just emerging. The memory device is almost certainly made possible by a change of phase along

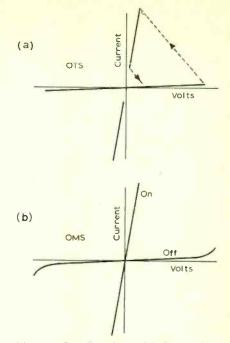


Fig. 5. Ovonic switches (a) the ovonic threshold switch: O.T.S. (b) the ovonic memory switch: O.M.S.

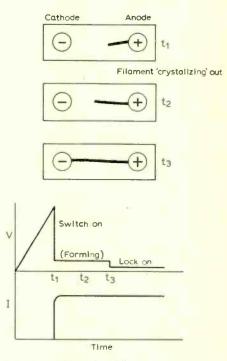


Fig. 6. Schematic of filament formation in memory device. When the voltage exceeds the threshold the current increases (switches on) but the filament takes time to form travelling from the anode. When fully formed the device resistance is lowest. This is termed locked on.

a filament; this change being induced by heating. The switching device is possibly tied up with a number of effects such as space charge, heating and contact conditions. But as one speaker at the conference said 'although we look through a glass, we look through a glass darkly'. This leaves lots of fascinating questions to be answered. Indeed switching and memory devices may not be the only uses that more knowledge about these materials could bring. It may be possible to develop specific glasses to absorb harmful wavelengths of radiation or indeed respond electrically to other wavelengths of light that existing technology does not permit. The biggest question for the industrialist is perhaps whether it will be worth the cost. The Cambridge conference probably ensured that firms with a current programme will maintain a holding programme of work. Then at least they have a hand in the field to pluck the flowers if they suddenly bloom in the spring. The lack of technological knowhow is unlikely to encourage many new firms to undertake their own research.

Acknowledgements

The author is indebted to Professor Sir Nevill Mott for allowing him to attend the conference at the last moment in spite of a full house.

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Amateur h.f. band

For many years, amateurs in Europe and Africa have been encouraged by the I.A.R.U. Region 1 Bureau to observe voluntarily an international "band plan" on the h.f. bands in order to reduce mutual interference between amateurs using different modes. While, at times, infringements of the plan can be heard (particularly the intrusion of phone operation into the c.w. segments), the plan has undoubtedly played a major role in maintaining orderly operation. The band plan was modified slightly at the Brussels I.A.R.U. Conference a few months ago, and is now as follows: 3.5 to 3.6MHz c.w. only; 3.6 to 3.8 c.w. and 'phone; 7.0 to 7.04 c.w. only; 7.04 to 7.1 c.w. and 'phone; 14.0 to 14.1 c.w. only; 14.1 to 14.35 c.w. and 'phone; 21.0 to 21.15 c.w. only; 21.15 to 21.45 c.w. and 'phone; 28.0 to 28.2 c.w. only; and 28.2 to 29.7 c.w. and 'phone. Radio teleprinter operation is recommended around 14.09 MHz.

Application Notes

Circuitry selected from device manufacturers' literature

Square-wave generator

0.001

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41-

The circuit given below operates over the following five frequency ranges: 2-20Hz, 20-200Hz, 200Hz-2kHz, 2-20kHz, and > 20kHz. P_2 is a coarse frequency control and P_1 is a fine frequency control

13k

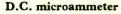
+15V

4A709

BAY7

Output

50k Ω ; ×100, $R_f = 101k \Omega$ Typical drift is quoted as $15\mu V/^{\circ}$ C. Extracted from Plessey Technical Communication No. 7.

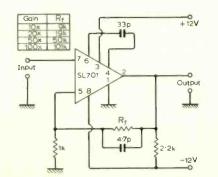


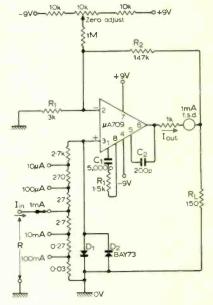
Below is the circuit of a low voltage-drop microammeter which will give an accuracy of 1% at ambient temperature if a good quality meter and accurate resistor values are used. Variation

which operates by varying the hysteresis cycle. Extracted from "The Application of Linear Microcircuits", Vol.1, SGS Ltd.

General-Purpose Amplifier

The gain of this amplifier is set by the resistor R_f : 10, $R_f = 9 \text{ k} \Omega$, $\times 20$, $R_f = 19 \text{ k} \Omega$, $\times 50$, $R_f =$





of accuracy with temperature is given as 0.2% °C. Extracted from: "The Application of Linear Microcircuits", Vol.1, SGS Ltd.

Magnetoresistance and its Application

Mean-square ammeter and d.c. transformer

by B. E. Jones*, M.Sc., Ph.D.

The magnetoresistance effect and the related Hall effect displayed by a semiconductor under a magnetic field have become of interest in recent years with the advent of extremely high-mobility materials.^{1,2} Both effects arise from the action of the externallyapplied magnetic field in producing a sideways deflection of the mobile carriers taking part in the conduction process.

The magnetoresistance effect is a phenomenon in which the resistivity of a semiconductor material is considerably increased by a magnetic field whenever the carrier mobility has a large value, for example in indium antimonide or indium arsenide intermetallic compound semiconductors. It has been shown that the total resistance of a rectangular specimen of such a semiconductor shows a square-law increase at small magnetic fields (up to about 0.3 T⁺) and a linear increase at small magnetic fields (Fig. I and Table I). The magnitude and characteristics of the effect depend largely on the geometry as well as the material itself.³ For suitably designed components the typical dependence of R_B/R_0 on magnetic flux density *B* applies to frequencies well into the gigahertz range. Wafer-shaped configurations exhibit their greatest sensitivity with the field perpendicular to the plane surface.

The electrical properties of a semiconductor are usually sensitive to temperature, and the magnetoresistance effect is no exception. Resistivity usually decreases with temperature and the larger the semiconductor surface area, the smaller the temperature coefficient In Table 1, one device has a temperature coefficient of $-1.8 \ \%/^{\circ}C$, while a slightly bigger and less sensitive device has a smaller temperature coefficient of $-0.12 \ \%/^{\circ}C$, both figures at B = 0, and temperature 25°C (these coefficients increase with B).

The linear magnetoresistance effect at relatively high magnetic fields has been used to produce a multiplying action, particularly for power measurement from direct current to microwaves.³ A transducer for displacement measurement based on the effect gave a large output of 5 V d.c. at 500 μ m displacement, without using electronic amplifiers, over a working temperature range -320° F to $+200^{\circ}$ F.² A magnetoresistance can obviously be used for magnetic-field measurement, particularly weak fields, and has been employed as a modulator of d.c. currents and voltages, as a contactless variable resistor and been applied to a brushless d.c. motor.¹

Two further application of magnetoresistances are described below. In the first case use is made of the square-law characteristic at low values of magnetic-flux density, to produce a simple clip-on mean-square ammeter (0-25 A) of very low input impedance suitable for measuring practically any current waveform. In the

[°]Electrical Engineering Laboratory of Manchester University

T is the symbol of the tesla, the SI unit for magnetic flux density (= 10' gauss)

Table 1

Characteristics of Magnetoresistance Elements

Туре	R₀ at 25°C (Ω)	R_B/R_e factor $B = \pm 0.3$ T, 25°C	Temp. coeff. at $B = 0, 25^{\circ}C ({^{\circ}}_{0}/{^{\circ}C})$
FP28D470	470	3	-1·8
FP17L100	100	1-9	-0·12

Note. The magnetoresistance elements are indium antimonide type made by Siemans & Halske A.G. and are obtainable in the U.K. from R. H. Cole Electronics Ltd., 7–15 Lansdowne Road, Croydon, CR9 2HB.

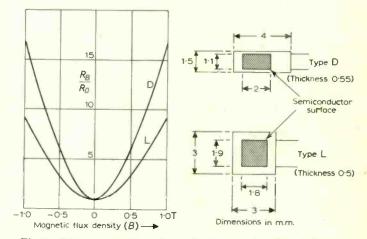


Fig. 1. Magnetoresistance relation R_B/R_0 as a function of the magnetic flux density B for two commercial devices.

second case, magnetoresistance is used as a magnetic-flux error detector in a feedback circuit to provide a simple d.c. transformer (0-1 A) for clip-on purposes.

Mean-square ammeter

The circuit used to test the mean-square ammeter scheme employing magnetoresistance is shown in Fig. 2. A gap is cut in one half of the small ferrite ring sufficient to take magnetoresistance R_1 mounted with silicon grease on a thin copper plate (the airgap also linearises the relation between magnetic flux and cable current; actually flux density in the ring

$$B=\frac{1\cdot 26\ I}{(l_t/\mu+l_g)}\,\mu T,$$

where I is cable current, l_i and l_g are mean length of magnetic circuit and airgap width respectively in metres, and μ is the ferrite magnetic relative permeability). To allow temperature compensation a second magnetoresistance R_2 is similarly mounted on the plate, but situated outside the magnetic circuit. To provide a meter deflection linearly related to change in R_1 caused by flux changes, both R_1 and R_2 are connected in a bridge circuit whose other two arms are current sources and the choice of out-ofbalance detector (resistance R_d) depends on accuracy and ruggedness required. The current sources are provided by two silicon transistors in a long-tailed pair arrangement with a well-defined voltage on the bases. With flux at zero (I = 0), potentiometer P_1 can be adjusted to balance the bridge.

For low flux densities $R_1 = R_0 + KI^2$, and for the case I = isin ωt , it has been shown⁴ that, considering only first-order terms, the mean detector current is given by the expression

$$I_{d} \approx \frac{KI_{1}}{R_{d} + R_{2} + R_{0}} \cdot \frac{i^{2}}{2} \left[1 - \frac{3K}{2} \cdot \frac{i^{2}}{2} / (R_{d} + R_{2} + R_{0}) \right]$$

assuming $I_1 = I_2 R_2 / R_0$, where I_1 and I_2 are the collector currents of transistors T_{r_1} and T_{r_2} respectively. It is evident that the detector current is proportional to the mean-square current $i^2/2$ in the



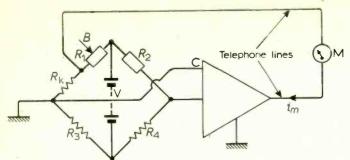


Fig. 2. Mean-square ammeter circuit. $F_1 = type A_1$ manganese zinc ferrite (Ferroxcube toroid FX 1322) (i.d. = 6.35 mm, $l_l = 2.76$ cm, $l_g = 0.59$ mm, $\mu_{min} = 100$). T_{r_1} , $T_{r_2} = type C444$ transistors (SGS Fairchild). R_1 , $R_2 = type$ FP28D470 magnetoresistances (see Fig. 1 and Table 1).

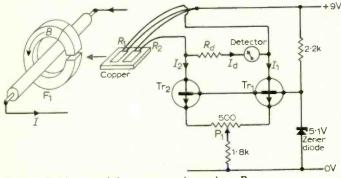


Fig. 3. Bridge containing compensating resistor R_k.

cable, provided the term after the minus sign in the expression for I_d is small. With $R_2 \simeq R_0 \simeq 470 \ \Omega$, $K \simeq 0.04 \ \Omega/A^2$, so that for 2% accuracy when $i^2/2 = 25 \ A$, R_d should have a value of about 900 Ω and $I_d \simeq 15 \ \mu$ A if $I_1 \simeq 1 \ m$ A (this value for I_1 restricts magnetoresistance dissipation to about $1/2 \ m$ W).

To check accuracy and linearity of the magnetic circuit a high impedance d.c. galvanometer ($R_d = 12 \ k\Omega$, $2 \ \mu A$ f.s.d.) was used as detector in the circuit of Fig. 2. There was less than 1% inaccuracy in the overall square-law characteristic for full-scale deflection when $I_{r.m.s.} = 25$ A at 50 Hz (the maximum flux density was only 0.07 T). The impedance of the ammeter is inductive, given by that of a single-turn on the magnetic core. With the core of Fig. 2, at 50 Hz, $X_m \simeq 6 \ \mu \Omega$.

Alternative arrangement

If it was desired to use a cheap standard rugged moving-coil meter as indicator at some distance from the point of measurement, the arrangement of Fig. 3 could be used. In this, R_k is a fixed compensating resistor in the branch of the bridge which contains the active element R_1 ; R_k is also connected to the output circuit of the d.c. amplifier. The input of this amplifier is connected to the output of the bridge and the amplifier output current i_m passes through R_k .

When a cable current (I) is present, the bridge is unbalanced, a voltage e is produced at the amplifier input and generates the current i_m . This current causes a compensating voltage at the terminals of R_k which is opposed to the error voltage generated by the change of resistance R_1 . The current i_m increases until these two voltages balance. It has been shown⁴ that

$$i_m = VKI^2/2R_k(R_k + R_0)$$

provided

$$2(R_k + R_0) + KI^2 / A \ll R_k (R_k + R_0)$$

 $R_1 = R_0 + KI^2$, V is the battery voltage across the bridge and $A = i_m/e$ is the amplifier gain. Thus i_m is independent of A and linearly related to I^2 so that a milliameter M will indicate the mean-square of the cable current. It is clear that R_k can be used as a range change control.

D.C. transformer

The alternating current transformer employs a large number of turns to keep small the magnetizing current required to produce the magnetic flux that opposes the flux produced by the current to be

Wireless World, January 1970

measured. Thus overall the current ratio of the transformer is accurately defined in terms of the turns ratio of the windings. The principle of minimum flux change has the advantage of operating the magnetic core material in a small fixed region of its magnetic characteristic, so that nonlinearity in this characteristic has negligible effect on performance. The principle has been used for direct-current measurement by employing magnetoresistance to measure d.c. flux.

The basic circuit employed to test the idea is shown in Fig. 4. A square ferromagnetic circuit containing an airgap and winding surrounds the insulated cable whose direct current is to be measured. Two magnetoresistances $(R_e' \text{ and } R_e)$ are attached with silicon grease to a thin copper plate to equalize their temperatures. They are connected in series and driven by a diode low-voltage source (V_d) to restrict dissipation. The active resistance R_e' is in the airgap, while the temperature compensating resistance R_e remains outside the gap. Because of the square-law characteristic of R_e' it is necessary to operate it at a constant bias flux density (B_k) , and this is produced by a stable fixed current (I_k) in the feedback winding (W).

If a direct current (I) occurs in the cable, a change of flux will occur in the magnetic circuit, R_e' resistance will change, as will the voltage (V) at the connecting point of R_e' and R_e . This voltage on being amplified by A and applied to resistance R_m , will produce a current (I_f) in the feedback winding (W) so as to produce a flux in the magnetic circuit to oppose the original flux produced by I. If the gain in the flux detector circuit and the voltage gain (A) are high, then the resultant flux change in the magnetic circuit will be very small, and I and I_f will be simply connected by the expression $I = NI_f$, where N is the number of turns of the feedback winding. The current I_f can be measured by means of a d.c. ammeter in series with the output of the amplifier.

The integrated amplifier has high gain (about 45,000) and produces a noise voltage of about 0.5μ V referred to its input, so it is necessary to use a 0.50 μ A meter in a low-pass filter circuit. When I = 0, zero meter deflection is obtained by adjustment of a stable offset current I_{0} .

It has been shown⁴ that I and I_f are in fact related by the expression

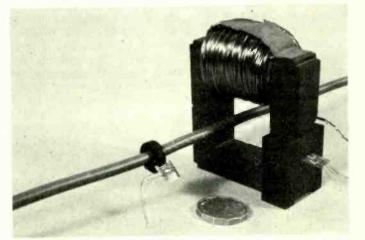
$$\frac{I_f}{I} = \frac{1}{N} \frac{1}{[1 + 1/K]}$$

where

$$K = \frac{NAR_e PB_k}{V_d R_m R_e'(l_l/\mu + l_g) \times 397} ,$$

the open-loop gain, P is the dissipation in R_e' , l_i and l_g are mean magnetic-circuit length and airgap width respectively, and μ is the ferrite magnetic relative permeability. With values N = 500, $A \simeq 45,000$, $R_e \simeq R_e' \simeq 100 \ \Omega$, $P \simeq 1 \ \text{mW}$, $B_k \simeq 0.06 \ \text{T}$, $V_d \simeq 0.7 \ \text{V}$, $R_m \simeq 500 \ \Omega$, $l_i = 18.7 \ \text{cm}$, $l_g = 0.89 \ \text{mm}$ and $\mu \simeq 1,000$, K has a value of about 10.

The amplifier gain fall-off is arranged to start at about 20 Hz



The magnetic circuits for the mean-square ammeter (left) and the d.c. transformer (right) are shown here. The magnetoresistances are also visible, but in operation one would be in the air gap.

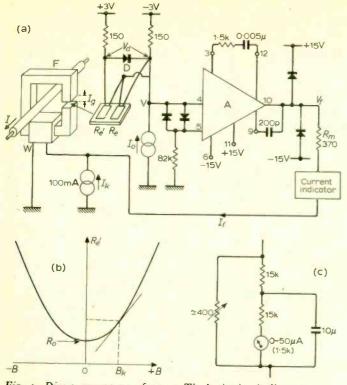


Fig. 4. Direct-current transformer. The basic circuit diagram (a) has R_e' , $R_e = type FP_{17}L_{100}$ magnetoresistances (see Fig. 1 and Table 1). All diodes = silicon, type ZS72 (Ferranti). $F = manganese-zinc ferrite, grade A5 (l_g = 0.89 mm,$ $<math>l_l = 18.7 cm, \mu = 1,000$ (Ferroxcube U-cores FX1795). W = 500 turn (N) winding of 24 s.w.g. $A = 45,000 \pm 50\%$ (MC1709CP integrated amplifier, Motorola).

to reduce noise and avoid instability problems. With full-scale deflection on the meter when I = 1 A, linearity is much better than 1% above I = 0.2 A; below this value amplifier nonlinearity at low signal levels has some effect. Wide variation in the cable diameter and position of cables in the magnetic circuit has no noticeable effect on readings. The input impedance of the transformer is very low by virtue of the effective high reluctance of the magnetic circuit. The current indicator could be at a considerable distance from the rest of the circuits.

Temperature effects

In both the experimental circuits utilizing magnetoresistances described above, the limit to current sensitivity and measurement accuracy is primarily fixed by drift, due to amplifier drift and magnetoresistance temperature dependence. Low temperature coefficient magnetoresistances, operated with minimum selfheating in compensating balanced arrangements should be used, and if necessary further temperature compensation can be achieved by a series thermistor or a parallel metal resistor. Amplifier drift can be reduced by using a d.c. chopper amplifier arrangement.

The experimental circuits described indicate two further useful applications of magnetoresistances. Both circuits are relatively simple and are useful for measuring currents in insulated cables by clipping a measuring head on to the cable. The mean-square ammeter is suitable for measuring practically any current waveform, while the direct-current transformer will measure low-frequency currents (for example, less than 10 Hz) where ordinary current transformers are inadequate.

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Weiss, H.; "Die feldplatte—ein neues magnetisch steuerbares halbleiterbauelement", Solid State Electronics, 1966, Vol. 9, pp. 443-451.

² Yuan, L. T.; "Magnetoresistive transducer", Solid State Electronics, 1966, Vol. 9, pp. 497-502.

^a Kataoka, S.; "Multiplying action of the magnetoresistance effect in semiconductors and its application to power measurements", *Proc. I.E.E.*, 1964, Vol. III, No. 11, pp. 1937-1947.

⁴ Jones, B. E.; Ph.D. thesis, Manchester University, 1969.

Announcements

"U.H.F./S.H.F. Techniques" is the title of a course of six evening lectures to be held at Norwood Technical College, Knight's Hill, London S.E.27, commencing February 3rd. Fee 15s.

M.E.C.-Electrosil Merger. Miniature Electronic Components Ltd, of Woking, has been merged with the Electrosil Group and all future enquiries and orders should be placed with Electrosil Ltd, P.O. Box 37, Pallion, Sunderland, Co. Durham.

Compat Telecommunications, a wholly owned subsidiary of Compat Corporation of New York, has established offices in Woolmead House, Woolmead, Farnham, Surrey, to handle all of the business of its parent company outside the U.S.A. Compat are manufacturers of computer-controlled data communications equipment.

Radiatron Components Ltd has been formed to operate in association with Radiatron Ltd and to deal with a wide range of components. It will handle the Elma range of collet knobs, stud switches, Elmaset instrument cases and readout counters. Both companies will operate from 76 Crown Road, Twickenham, Middx.

Racal Electronics and Kelvin Hughes have agreed to work in partnership on a range of h.f., s.s.b. marine radiotelephones. Racal have a contract from Kelvin Hughes for the design and manufacture; Kelvin Hughes the world-wide marketing of the products.

Industrial Control Systems Ltd have moved to 78-90 Clarke Road, Northampton. (Tel: Northampton 32417).

The Crawley offices and laboratories of **Pye Unicam** Ltd have been transferred to the company's head office at York Street, Cambridge, CB1 2PX.

Microwave & Electronic Systems Ltd, of Midlothian, Scotland, has moved its sales office to 66 Tilehurst Road, Reading, Berks. (Tel: Reading 581937/8.)

G. A. Stanley Palmer Ltd have been appointed sole U.K. agents for the range of miniature electrolytic aluminium capacitors manufactured by the International Electronics Corporation of Long Island, N.Y.

Hayden Laboratories Ltd, East House, Chiltern Avenue, Amersham, Bucks, have been appointed exclusive U.K. agents for **Spinner GmbH**, of Munich, W. Germany, manufacturers of radio frequency connectors, directional couplers and other specialized items associated with radio-frequency cables and waveguides.

Nobel Electronics, of Welling, Kent has signed a three-year agreement as sole **U.K. and European agents** for Plastic Capacitors Ltd, Maydown, Co. Londonderry, N. Ireland.

Montclair Electronics Inc, of New York, have appointed G. A. Stanley Palmer Ltd as sole U.K. agents for a range of magnetic reed relays and switches from the General Reed Company.

Low-distortion Bias and Erase Oscillator

Evolving a current switching design to give a predictable and stable output level and with no trimming requirement for low distortion

by D. Griffiths, Ph.D.

The design of a good bias and erase oscillator for a tape recorder is not easy. For stable biasing a constant output voltage is required, yet any limiting action in the oscillator must not be allowed to distort the sine-wave drive since this would increase the background tape noise. The circuit should be efficient so that the least expensive semiconductors can be used and, ideally, it ought to be designed to work straight off without any complex setting-up procedures, especially those required to minimize the distortion.

The oscillator was required to operate a Ferrograph Series 6 tape deck but using the procedure outlined below it should be possible to alter the component values to make it suit almost any other recorder.

Specification

The Ferrograph handbook gives the inductance of the two-track FE16 erase head as 1.5mH (per track), requiring 80mA at 27-30V and 68kHz; the record head only requires about 5mA at 15V. The power required from the drive circuit is fortunately not $0.08 \times 30 = 2.4$ watt but depends only on the losses in the heads; a perfect inductor can not have a net dissipation of energy. Since the erase head has a mu-metal core it is only useful to measure its loss under actual working conditions, for it is quite hopeless to try to extrapolate data on iron-cored inductors.

In the absence of a suitable measuring bridge the losses in this head were assessed by observing their damping effect on a tuned circuit resonating at 68kHz with 30V across it.

The inductance of this test circuit should be less than, say, one fifth of that of the tape head so that the resonant frequency is not too greatly changed by the extra head inductance, but the waveform need only be roughly sinusoidal.

It was found that each winding of the FE16 head introduced the same damping effect on the test circuit as did a $3.3k\Omega$ resistor. Under operating conditions in a parallel resonant circuit the head could thus be thought of as a perfect 1.5mH inductor in parallel with a $3.3k\Omega$

resistor, resulting in power dissipation of about 0.25 watt.

At this point one has to face firmly the problem of ensuring a constant output voltage. The necessary limiting action can use the "curvature of the characteristics" of the active circuit elements but designability is sacrificed and well stabilized power supplies are required to maintain operation in the critical region. Some form of a.g.c. could be employed but amplitude overshoot at switch on must be avoided. This is also a problem with thermistor stabilization. The alternative scheme chosen here is to send constant current pulses of a suitable shape through a tuned circuit coupled to the tape heads and rely on a reasonable Q value to reduce the harmonics sufficiently. This filtering is more effective than one might imagine since the harmonic amplitudes decrease with increasing order, while the attenuation of the tuned circuit also rapidly increases with rising frequency. With transistors or valves it is a simple matter to generate well enough regulated driving pulses for this application but the feasibility of the scheme depends entirely on maintaining a good Q in the filter despite the loading of the losses in the tape heads.

Current switching

Ideally the reader should now turn up Wireless World for November and December 1962 to an article by R. C. Foss and M. F. Sizmur which gives an admirably lucid account of current

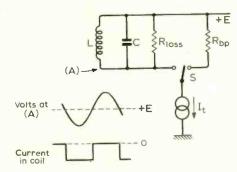


Fig. 1. Principle of a current switched LC oscillator.

switching sine wave oscillators. Their first diagram is reproduced here as Fig. 1 and is a good starting point. When the current generator is first connected to the tuned circuit the voltage at point (A) will swing below +E with a period governed by the resonant frequency. The size of this swing depends on the losses in the LC circuit, as well as the magnitude of the drive current. Eventually the voltage at (A) swings back up towards +E, even with the constant current generator still connected. When the voltage across the resonant circuit is zero (i.e. when the point (A) is at +E again) we choose to switch the current supply into a bypass resistor R_{bp} . The point (A) then continues its upward voltage swing and but for the losses this would take it as far positive above +E as it had been below +E half a cycle earlier; the current generator has infinite output impedance and cannot load the LC circuit. When the voltage at (A) eventually falls to +Eagain we reconnect the current supply to the tuned circuit and the cycle repeats.

As a step to a practical realization Fig. 2 shows the next stage of complication and is also from Foss & Sizmur's article. Here the tail current I_t is alternately switched between the transistor and the diode by the action of the voltage induced in the base winding N_b which is coupled to the tuned circuit. As indicated, a phase reversing connection is necessary so that when the point (A) is below +E the base end of N_b is positive with respect to ground and the transistor conducts as required. But for this base winding voltage and

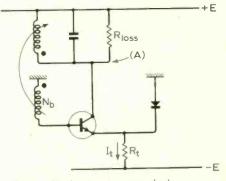


Fig. 2. Tail current is switched between the transistor and the diode.

 V_{be} , the tail current driving the LC resonator would be E/R_t . The tuned circuit is only lightly damped by the transistor as the collector is a high impedance point.

In Fig. 3 the diode is replaced by another transistor with its collector connected to a similar *LCR* circuit as that on the left hand side. Except for the brief instant of current changeover, the tail current flows only through either Tr_1 or Tr_2 . Whichever transistor is off will have its collector above +E while the other collector is equally below +E. That is, the voltages at (A) and (B) seesaw about the positive supply rail.

It is now only a small step to the final arrangement shown in Fig. 4. A single tuning capacitor is employed across a centre tapped inductor which has an additional winding to provide the required bias and erase voltages. As only a positive supply rail was available, the bottom of the tail resistor is connected to ground and the centre tap of the base winding supplied with a suitable potential stabilized with a zener diode. A pair of plastic encapsulated transistors is used for each switch to give the collector dissipation required when both erase heads are simultaneously connected in two channel operation. The 15Ω emitter resistors help to equalize the current in each pair and are useful inspection points at which to observe the individual current waveforms. The $2 \cdot 2k \Omega$ base resistors are a personal whim to reduce possible excessive base currents when trying out the prototype.

We choose to make the reference voltage defining the tail current about 2.5Vabove common. There are two reasons for using such a small value. First, that the collector voltages can have a large excursion which will entail a lower stepup ratio to achieve the desired output volts and hence a lower reflected loss from the heads, giving a better Q factor in the filter. The second reason is connected with reducing the output distortion, as discussed later.

To see if the negative going excursions are bottoming the transistors one does not check the voltage waveform at the collectors (!) since the flywheel effect of the high Q tuned circuit dominates the response; it is more useful to examine the tail voltage across R_t as shown in Fig. 5 and look for the 'dents' indicated. It must be remembered that during the off half cycle the transistor experiences a maximum collector voltage equal to the supply plus the amplitude of the downward swing. 12V r.m.s. is about the maximum reasonable collector excursion with the circuit values shown in Fig. 4. The maximum collector voltage is thus about 22 + (12×1.4) V. Even allowing for a peak emitter voltage of 3V (see Fig. 5), this uncomfortably exceeds the maximum recommended $V_{ce} = 30V$ of the 2N3704 transistors used in the prototype.

If the tail current is assumed to be constant during each cycle, the collector dissipation can be easily calculated once the mean collector-emitter voltage is known over the conducting half cycle. Since the average value of a halfsinewave is 0.64 times its peak value and if the collector peak swing is $12 \times \sqrt{2}$ volts, then the collectors are on average $12 \times \sqrt{2} \times 0.64 = 11V$ below 22Vwhen conducting, i.e. IIV. When operating, an Avo indicated 2.6V d.c. across the tail resistor, giving a tail current of 52mA. On average, $V_{cc} = 11 - 2.6 = 8.4 V$ and if the current is equally shared between each pair of transistors, the mean collector dissipation is $8.4 \times 0.052 \times 0.5 \times 0.5 \approx$ 110mW, with the second factor of 0.5 arising from the on-off time ratio.

If the two erase tracks are in use together, the extra drive can be achieved by suitably reducing the tail resistor and there is still a reasonable margin of collector dissipation in hand. This circuit is not very efficient in terms of power consumption; the four transistors dissipate ≈ 450 mW to overcome a head loss of 250mW.

Although a good L/C ratio is needed to minimize losses in the tuned circuit primary, the maximum allowable primary inductance is set by the Q value which has to be maintained in spite of the damping effect of the tape head losses. With a I:2 step-up ratio between primary and secondary the equivalent loss resistance of $3\cdot 3k \Omega$ looks like 820Ω across the primary circuit. If the circuit shows a Q value of Q_f at resonant frequency ω_f , the dynamic resistance of the circuit is $Q_f \omega_f L$. Clearly, in this case, we need $Q_f \omega_f L \leq 820 \Omega$.

Now how much Q_f is needed? Foss and Sizmur show that with square current pulses the ratio of the n^{th} harmonic voltage V_n to the fundamental V_1 is given by:

$$\frac{V_n}{V_1} = \frac{1}{(n^2 - 1) \cdot Q_f}$$

As a square wave can only generate odd harmonics, the third order one will be the principal component and with Q = 10 its amplitude will be 1.25% of the fundamental. Since it is planned to use something a little less brutal than square driving pulses, this Q value should suffice. For operation at 68kHz this fixes L at 0.25mH. A 30mm diameter ferrite pot core, Mullard LA 2202, was used (with a permeability of 63) giving 1mH for 60 turns. The primary was wound in bifilar fashion to give 30 turns centre-tapped, with 60 turns on the secondary. 28 s.w.g. enamelled copper wire was used for both windings. The working flux density in this application is in the region of 50-100 gauss.

The tuning capacitor has to be larger than the value required to tune the 0.25mH primary inductance to 68kHz since the inductance of the tape head is reflected into the resonant circuit with a magnitude reduced by the square of the

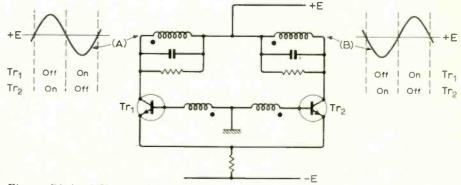
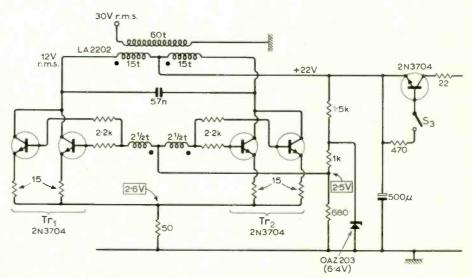
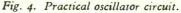


Fig. 3. Diode of Fig. 2 replaced by second transistor.



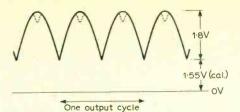


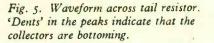
step-down turns ratio; the total inductance is thus lowered since "inductors in parallel add like resistors in parallel". The secondary winding itself does not behave as a separate inductor since the induction in it is solely determined by that needed to balance the primary applied voltage. If the a.c. voltage across a coil does not depend on the rate of current change through it, then it does not have inductive properties.

Although Ferrograph quote a nominal erase inductance of 1.5mH, their suggested operating point of 80mA and 30V r.m.s. indicates a working inductance more like o.9mH. As the Vinkor pot core was used without an adjuster, its inductance would be about 10% below nominal. Together with the slight contribution from the recording head, the total effective inductance would thus require about 51,000pF to tune it to 68kHz; in practice 10% tolerance 47,000pF and 10,000pF capacitors were used in parallel. If tuned filters are used as bias rejectors in the recording amplifiers, it will be important to keep the same bias frequency on single and two track operation. The extra inductance of the second head would lower the frequency further and a suitable extra tuning capacitor would have to be switched in.

It must be confessed that there is a little bit more complication in Fig. 4 than was admitted in earlier paragraphs and this concerns the shape of the current pulses. A square pulse with its sharp edges is obviously a rather poor approximation to the required output waveform and something a little more sinewave-like would ease the filtering problem. Now one cannot go to the limit and use an exactly sinewave current drive derived from the output waveform as there is then no limiting action, other than unintentional clipping and bottoming, etc. As a compromise we use a current pulse which is "partly square and partly sine". This is illustrated in Fig. 5 which shows the alternating voltage waveform across the 50Ω resistor, superimposed on the calculated d.c. level which could not be observed with the a.c. coupled scope available.

The squarewave part of the current waveform is developed by the long tail switching action; the sinewave part has a similar amplitude and is derived from the filtered output. This is achieved by giving the base windings a suitable number of turns so as to inject an appropriate amount of sinewave signal in series with the steady d.c. reference level. However, one must be careful in selecting the amplitude of this a.c. component otherwise the maximum reverse bias rating of the emitter-base junctions will be exceeded and extra protective diodes will be needed; the reverse emitter-base rating for the 2N3704 is given as 5 volts. It is important to recognize that the 'off' base junction sees both base windings in series generating the reverse voltage.





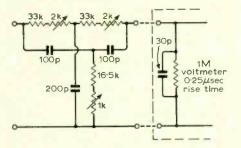


Fig. 7. Twin-T filter for 68kHz.

Fig. 6 shows the base of $Tr_2 + V_b$ above V_{ref} derived from the zener diode. But for the V_{be} of this 'on' transistor and the voltage drop across its 15Ω emitter resistor the top of the tail resistor would also be $+V_b$ above its d.c. level, taking the emitter of Tr_1 with it (in the positive going direction). Meanwhile the voltage $-V_b$ on the left hand base winding is holding the base of this transistor down $-V_b$ below V_{ref} .

The 1.55V d.c. level in Fig. 5 assumes a Vbe of 0.7V and allows for the two 15 Ω resistors in parallel. The mean level of the a.c. component is $1.8 \times 0.64 = 1.15$ V and thus an Avo on a d.c. range across the tail resistor should register (1.55 + 1.15) = 2.7V. This agrees well with the 2.6V observed. Evidently the steady current component is 32mA and the r.m.s. a.c. contribution is 25mA. In a Fourier representation of a square wave, the first harmonic has an amplitude of $4/\pi$ times the amplitude of the square wave. Adding these two contributions to the voltage developed across the dynamic resistance seen in the primary circuit, one can estimate the loss resistance at 570 Ω , corresponding to a working Q of 6—which is rather below the design figure.

The 500μ F reservoir capacitor in Fig. 4 ensures that the oscillations decay smoothly when the circuit is switched off, thus helping to keep the tape heads demagnetized. The decay time is 0.5-1s. Switch S₃ is controlled by the deck selector knob and operates via the series 2N3704 to minimize peak current through the switch contacts. The 22 Ω resistor ensures that the initial charging current of the 500μ F capacitor does not greatly exceed the maximum transistor current rating of 800mA.

Performance

After all that story, how does it do its job? The amplitude of the output

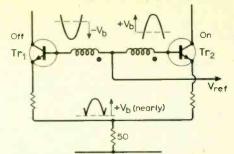


Fig. 6. How base-winding voltages add to build up reverse emitter-base voltage.

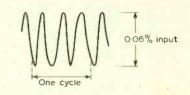


Fig. 8. Residual waveform at the base winding after passing through the twin-T filter.

slowly increases by 2-3% during the first half minute or so after switch on and this is probably due to heating of the transistors. A 40°C rise in junction temperature would lower V_{be} by about 90mV and thus increase the standing tail current. enough to account for this observed rise in output. This effect could be reduced by increasing the tail voltage, remembering to add diodes to protect the baseemitter junctions from excessive reverse voltages arising from the necessary accompanying increase in sine wave drive. It is doubtful if the present small change in biasing could possibly be detected by its effect on the recorded signal. Changes in the supply voltage only slightly affect the oscillator output as might be expected; a 10% reduction in supply potential reduced a 28V output by just under 1.5%.

In the absence of a wave analyser, a simple passive twin-T rejector was used to filter out the fundamental to see what was left. Fig. 7 shows the filter circuit, and Fig. 8 shows a sketch of the residual signal from the 16V bias at a recording head on single channel operation. The amplitude of this residual is 0.06% of the input level and appears to be largely 3rd harmonic as expected. A sine wave input of 3×68 kHz to the filter was attenuated by about 7dB as seen on the voltmeter, so it seems likely that these distortion products do not have an amplitude exceeding, say, 0.2% of the fundamental. This seems quite satisfactory and shows that the idea of using current pulses with the "edges rounded off" does indeed greatly reduce the output distortion while still retaining an adequate stability of output level. With a working Q of 6, square drive pulses would have given a 3rd harmonic component an order of magnitude greater at 2% of full output. Judged audibly, the tape hiss is very low and BASF double play tape on the Ferrograph appears to give a peak signal-to-hiss ratio in the upper fifties of decibels.

Industrial Telemetry

Some recent supervisory and control schemes

by R. E. Young

The early 1960s saw major developments take place in industrial telemetry¹, largely as the result of the wide introduction of solid-state equipment and digital techniques. Rapid expansion then followed in step with the accelerating demand for these forms of automation backed by the extremely high reliability that they had been shown to give.

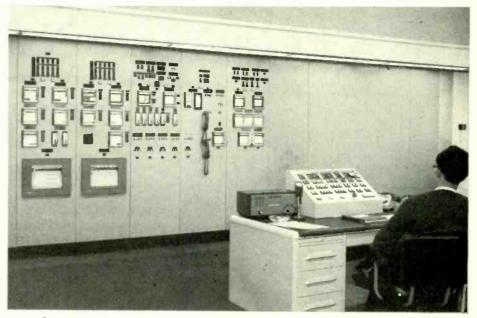
Much of this expansion occurred in the public utility field, authorities being strongly influenced in their policy by the almost overwhelming growth of "service" distribution networks, e.g. for electricity and water, and the increasing cost of manning them in the conventional way.

In general, the economy shown by the adoption of supervisory telemetry methods increases with the size and complexity of the project. Furthermore, the most favourable conditions for setting up such remote control systems are usually found with high concentrations of population and industry.

Thus in a large scale installation for electricity supply in the Far East control is exercised over the distribution network for the urban area of Kuala Lumpur², three bulk supply stations and a total of fifteen substations being covered by the first phase of the scheme.

Recently commissioned, this is a classical digital supervisory system with time-division multiplexed telemetering and telecontrol, and employing the interrogation /reply, or responder, techniques which are used almost exclusively for this work³. With time-division operation, each information source is scanned in turn, and in these systems this is achieved by interrogating each source in terms of the unique (digital) address allocated to it.

Measurement or equivalent data points are grouped in blocks of addresses according to priority, so that the period which elapses between successive scans of a given point represents the "updating" time for its particular address block. With the exception of control functions, the various groups of addresses are interrogated in accordance with a pre-determined scan cycle, system programming being arranged to interlace these addresses within the overall scan period. Typically, an updating period of 11 seconds is realized for some 80



Control room at the Shell shore terminal at Bacton for their North Seas gas project.

measurement addresses; while alarm indication (e.g. for abnormal transformer oil temperature), carrying more urgency, is given a block updating time of 5 seconds.

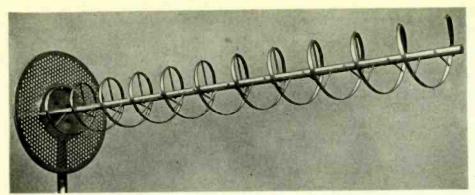
Control instructions are also sent out by interlacing, but this is done by interrupting the routine scan cycle and thereby extending it for the additional time-sharing to take place. The same system word format is used for the control "way" addresses as for monitoring, and also in both cases the encoded replies sent back from the outstation follow the same pattern.

Obviously, measurements and monitoring information generally must be presented as far as is possible without any likelihood of there being any ambiguity or misinterpretation of the intelligence. It is of interest that the network electrical measurements, viz. voltage, power, reactive power and current, are displayed in analogue form on conventional d.c. moving-coil meters. This involves the use of an individual digital-analogue convertor for each meter as the incoming signals are handled digitally throughout the logic system. It may be taken that the factor of additional cost is considered more than offset by the advantages accruing from working with a familiar form of display. Also it may be noted that the eight-bit binary number d.a.c.s which are employed, and are of the "successive approximation" type, act as information stores over the measurement updating period.

Communication between master and outstations is by "four-wire" working using special modems* developed to give maximum speed data transmission over nominal 3kHz bandwidth circuits. These links are set up either in existing pilot cables where suitable spare cores are available or in the main communication cables run along the power network routes. Protection against extraneously induced high voltages is provided at all station line terminations by lightning arrestors and isolation transformers.

The other main area of application for supervisory telemetry systems is the monitoring and control of oil and natural gas pipe line schemes. Here inaccessibility of wellheads and pumping stations is a

modulator-demodulator (unit), "four-wire" working demands two conductor pairs, one for "go", one for "return".



Helical aerial supplied by C & S Antennas Ltd for the North Sea gas radio links.

major driving force in the adoption of such schemes; these conditions, compared with those associated with most public utility networks, tend to impose more severe restrictions on the choices open to the telemetry system designer. Thus with projects such as the North Sea installations, the virtually inescapable use of radio links produces a "design constraint" which affects the whole system.

The same basic time-division digital techniques are employed, however, for these installations as for the public utility networks; and logic circuit blocks and address-reply methods are essentially the same. The telemetry installation for the B.P. group of wells in the West Sole field⁴ is typical of such practice for North Sea operations, the outstations on the project's three wellhead platforms-"A", "B" and "E1"-being under the supervisory control of the shore station at Easington, Yorks. As commissioned, the system capacity is 22 measurements, 120 monitor and alarm indications, together with 41 well control functions; the routine scan updating time is 25 seconds.

One of the main system operational requirements arising from its production control function is the calculation of mass flow for each of the eleven wellheads involved. These corrected values have to be obtained in terms of differential pressure type flow measurements and corresponding manifold pressures; and initially a study was made of using an individual analogue computer at each wellhead. However, it proved possible to centre this function on a single digital computer which is fed with the data in digitized form, and which gives flow rates as a 3-digit numerical indication up to a maximum per wellhead of 59.9 million cu.ft./day.

In general, measurements are displayed on the mimic control panel with three-digit representation for temperature and four for pressure, a "scaling" facility being incorporated in the transfer from binary code input to the decimal reading output. Accompanying this implied degree of resolution, a 12-bit format is used for both addresses and replies, each carrying three additional parity bits for error checking. With the system parameters obtaining in this case, the address/reply cycle time becomes 570 milliseconds. For transmission over the radio links, the address and reply pulse trains are converted in a frequency shift keying modem to a "tone" input for the transmitter—2.3kHz for binary '0' and 2.7kHz for binary '1'.

The u.h.f. radio link scheme adopted for this project operates in the 460-MHz band and inevitably invites comparison with the offshore wellhead control scheme at Das Island in the Arabian Gulf (Umm Shaif oilfield). Described originally in 1964⁵, this employs a microwave, 3cm ('X' band) link based on a commercially available transmitter magnetron with a rated peak power output of 2.5 kW in the centre of the band. The main point of interest in the present context is that in this earlier scheme a single transmitter is used with radiation from a "cheese" reflector giving a half power beamwidth of about 40° in azimuth to cover the fan-like sector in which the wellheads are grouped.

In contrast the North Sea u.h.f. system utilizes a two stage "hand-on" arrangement for signal transmission between wellhead platforms. The primary link is established between the Easington master and the outstation on platform A, working between this platform and both platforms B and E1 being on a "broadcast" as distinct from a beamed mode. Thus the transmitters on B and E1 operate on a shared frequency, and, in order to avoid radiating together, "come on the air" only when their own plant addresses are received.

Helical aerials mounted on 200-ft towers are used for transmission and reception on shore and at the platforms. The 11-turn helical elements, made by C & S Antennas Ltd, have a rated beamwidth of 30° to half power points with a v.s.w.r. of 1.5 over the operating bandwidth of 400-500 MHz.

The radio link equipment itself is solid-state throughout; Standard Telephones and Cables type HTR20 f.m. transmitter /receivers being employed with a nominal transmitter power output of 5 watts. This output is obtained from a varactor tuned to act as a trebler stage fed at 133.3-163.3MHz from two preceding trebler stages which have a modulated input at 14.8-17.8MHz. This latter input is obtained from a two-stage phase modulator with crystal oscillator reference drive. Two stages of amplification are interposed between the modulator and the first set of treblers which is followed by three more stages of amplification to give the input to the final varactor trebler. A tunable bandpass filter is placed in the output from this varactor stage to act as a harmonic suppressor.

In the double superheterodyne receiver the first mixer is preceded by a two section bandpass filter and two stages of r.f. amplification. The bandpass filter is largely responsible for the degree of r.f. selectivity and second channel rejection achieved. A single crystal oscillator feeds both mixer stages, the higher local oscillator frequency required for the first mixer being obtained by multiplication by six (doubler followed by trebler). This avoids the production of spurious beats which is possible with two separate oscillators, "spurious responses" being given as below -80dB. Intermediate frequencies are 70MHz and 10.7MHz with an initial local oscillator frequency of 55-68MHz, fed to the second mixer, and multiplied to 330-408MHz for the first mixer. The output of the second mixer, nominally at 10.7MHz, is fed into a crystal filter to give selectivity at this frequency and thence to a wideband amplifier which provides the input to the limiter and discriminator stages. Both the wideband amplifier and these stages are constructed as linear integrated circuits. Performance criteria are based on a minimum acceptable signal /noise ratio of 20dB; while, for the individual radiation requirement of the broadcast mode, carrier "on" switching time is given as not more than 1mS.

"Telegrid" master programming control System

As already indicated for the two schemes described, system working speed (data handling speed) is kept relatively low, i.e. the equivalent of a narrow-band telephony channel is generally employed for communication in present generation supervisory telemetry projects. Nevertheless, these communication links must be highly "secure" and, equally important, economic in the full sense of the word; and with the ever growing demand for telecommunication channels, this latter condition is becoming increasingly difficult to meet. This difficulty is encountered whether line or radio working is adopted because of limited capacity-scarcity of installed cables, particularly in built-up areas, and on the radio side, severely restricted channel allocations for such applications.

It is with this background that the Telegrid proposals were put forward as a means of "multiplying" the number of existing communication channels by what may be called supra-multiplexing under the control of a master programming source. The scheme, proposed by G. S. Kermack, managing director of Serck Controls, makes specific communication channels available to users, grouped on a network basis, in accordance with a time sharing schedule held in sequence by narrow band synchronizing signals. Planning of such a scheme would have to be on a national scale, although operation might be on a regional basis within the national framework.

Wireless World, January 1970

In one suggested embodiment of the scheme (Figs. 1 and 2), four networks are time multiplexed under the control of broadcast synchronizing signals. Network allocation, as shown, would be electricity, gas and water for a distribution group, together with an emergency or stand-by network available to take over from any one of the other three. Alternatively, network 4 could be utilized to give a low speed data transmission facility over a large area in the event of, say, major floods occurring.

The programming of these networks is carried out by a combination of imposed synchronization and delay timing. For this the networks are grouped into two pairs, with the first member of the pair taking the external synchronizing signals, and the second becoming operative after a predetermined time delay following the commencement of the first network scanning cycle or sub-programme.

The main technical feature of the system is the form of coded signals used for programming the networks. These signals are built up from "pips", i.e. short bursts, of tone which can be broadcast from a low-frequency (say 300kHz) transmitter to cover a regional area. In addition to their task of time division synchronization, these master programming signals perform two other functions:—

(i) Designation and identification of the network to be activated;

(ii) "Start" the individual network scanning cycle after receipt of the correct combination of signals.

The latter provision is achieved by arranging that the five pips must have been preceded by the six pips before the "five-pip" group is opened up, and conversely. If this sequence is not maintained owing to the absence of a signal or the presence of spurious signals, then the networks are not activated until the correct sequence is re-established, i.e. the system has been made to "fail-safe".

From the diagram it will be seen that guard spaces form part of the timing pattern. These take care of short-term variations and fault condition in individual sub-programmes.

Other developments employing these techniques can be envisaged, as, for example to arrange for each pip or burst of tone to contain a predetermined number of cycles, and, by counting at the

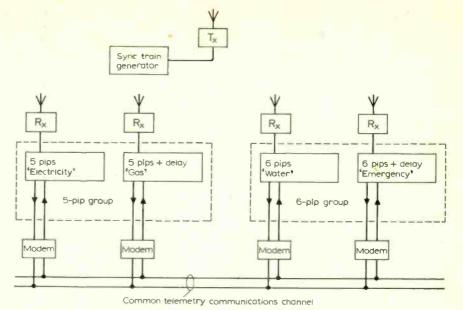


Fig. 1 "Telegrid" diagrammatic network arrangement.

receiving end, to obtain further complementary checking and possibly more precise synchronization.

One of the main advantages of the system is that with an accurately maintained pip (tone) frequency, say at 400 Hz, the signal extraction band-pass filters can be made extremely sharp and only a "crevasse" is required within the synchronizing channel transmission spectrum. Furthermore, with such band-pass filters (e.g. crystal or mechanical type) high rejection of spurious signals is obtained.

Television Link on Low Bandwidth Cable

Television, as a time-division system, is part of the telemetry family; and in presenting visually inaccessible conventional gauges and similar instruments fulfils a specific telemetering function. One of the main attractions of such presentation is that effectively there is no updating delay, and—often of more importance—rapid changes in quantities can be seen on analogue displays via a television link, whereas they are beyond the capability of the comparatively slow scanning telemetry system.

To speed up these telemetry scanning

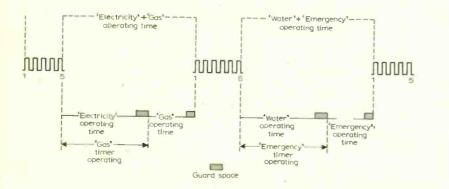


Fig. 2 Overall timing pattern of "Telegrid".

rates to give the equivalent of a television system, though theoretically possible, becomes prohibitive in cost. There are instances, therefore, when a television scheme provides the most economic way of tackling an unconventional instrumentation problem, this being much more marked when it can be used for other monitoring duties as well.

A variant of such a scheme is represented by the East Anglian Water Company's closed-circuit television installation at Lowestoft where the emphasis is on surveillance rather than instrument monitoring. The outstanding feature of this project is the video link. As far as is known, this is the first time that a link has provided operationally acceptable picture quality over a 5.2 mile (8.35 km) length of "telephone grade" cable without intermediate repeaters.

This link is of interest on two counts. The first is the potential offered by the equalization and allied techniques which have been developed and shown to be effective under these conditions for high-speed pulse transmission. This aspect bears directly on the problem of obtaining maximum data transmission speed on restricted bandwidth circuits, and also on the improvement in error rate produced by equalizer correction of signal distortion.

The second point is that compromise on picture standards had to be reached but that it proved possible to use a field rate of 50 per second instead of the much lower rate proposed at first in view of the "no-repeater" and other limitations. It was clear that a 405-line interlaced structure was the absolute maximum that could be attempted in terms of frequency and this had the advantage that comparatively low-cost U.K. standard camera and monitor equipment could be employed.

The S.T.C. ten-pair cable installed by the water company between its intake and borehole station at Belaugh and the Horning master control for both



Telemetry and television supervisory control position at the Horning master station of the East Anglian Water Company showing monitor picture as received over the low bandwidth cable link.

telemetry and television signal transmission is of the polythene insulated type with outside steel tape armouring acting as a screen. Diameter overall is some 22mm, individual conductors being of 0.9mm diameter. Conductor resistance is given as 44.2 ohms per mile at 15°C, with attenuation at audio frequencies of 1.20 dB/mile and crosstalk between pairs better than -80dB measured on site. Attenuation reaches a value of some 80dB down at 1.2MHz with an unequalized frequency response approximating to the form $1/\sqrt{1requency}$.

In the final solution, the video circuit was established as two conductor pairs diametrically opposite each other in the cable and connected in parallel. This was found marginally better than a single pair circuit; and at the output of the receiving end equalizing amplifier a uniform response within 2dB is obtained up to 1MHz. The overall response is about 6dB down at 1.2MHz with relatively sharp cut-off thereafter.

It should be noted that a contribution to improved high-frequency response is made by including pre-emphasis in the transmitting characteristic. This amounts to 10dB with a 3dB point at 200kHz. The necessary phase equalization is carried out at the receiving end, some 0.5 microsec. correction being given at 1MHz.

Finally a "crispener" is incorporated to sharpen up fast edges in the video waveform by speeding up their rise times in a non-linear network. The crispener unit embodies an input filter to extract the fast edges from the equalized video waveform for feeding to the non-linear system. This filter operates by signal subtraction referred to a wideband delay line to give a Gaussian type response. After leaving the non-linear network, the artificially sharpened edges are recombined with the original video waveform, the delay time in the filter being compensated by introducing a corresponding delay in the main video path.

The results obtained in respect of signal/noise ratio, better than "dusk" camera input level (10 foot-candles), can be ascribed to the precautions taken, e.g. with regard to common mode rejection (receiving head amplifier rejection ratio of 70dB), and to the maintenance of electrical balance about earth at the appropriate points in the sytem. The picture reproduction gains considerably from the crispening technique, although quite acceptable without it; a rise time of less than 800 nanosec being obtained on a 10kHz square wave as measured at the input to the crispener. The overall picture quality also benefits from the use of a monitor with a black level clamp.

Acknowledgement must be made to L. G. Davis, of Glenn Sound Services, who was responsible for the special television link equipment described, and to Serck Controls for supplying details of the two supervisory schemes covered in the first part of the article.

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We regret the delay in the publication of the reprint of the articles covering the Bailey 30-W and 20-W amplifiers and pre-amplifier. This is now available. For the convenience of new readers we give below the full list of W.W. reprints obtainable from the Trade Counter, Dorset House, Stamford Street, London S.E.1. Prices include postage and packing.

No. 1. High-fidelity Amplifiers by A. R. Bailey (Nov. and Dec. 1966, and May, June and Nov. 1968). Contains articles on 20- and 30-W amplifiers; a pre-amplifier; and an output transistor protection plus modifications and relevant correspondence. Price 5s.

No. 2. Stereo Decoder and Simulator by D. E. O'N. Waddington (Jan. and Oct. 1967). Describes the construction of a stereo decoder for positive or negative power supplies and contains details of an instrument for producing a stereo multiplex signal. Price 3s.

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Active Filters 6. Lead-Lag network and positive gain by F. E. J. Girling* and E. F. Good*

The well-known Sallen-and-Key low-pass and high-pass circuits provide two of the most useful building bricks for applications where only low or moderate values of Q factor are needed. They are practical examples of the second type of active system analysed in Part 4, a lead-lag or lag-lead network and positive gain, with input connections changed to give low-pass or highpass response as the case may be.

A notch (or zero) in the stop band is easily obtained by adding a parallel path. This gives a section with a characteristic useful in the realisation of a high-order filter as a cascade (or product) of factors.

Adaptations which give "tunedcircuit" response are also described.

The Sallen-and-Key circuit

The lead-lag network in a loop with positive gain, K, has been analysed in general terms in Part 4. This analysis can be applied to the Sallen-and-Key low-pass circuit, Fig. 1(a), by reference to Fig. 1(b), which shows the same circuit with the input V_1 shorted out and a floating generator V_2 introduced into the feedback path. If now the loop is supposed opened at X and the freed end of V_2 earthed, it can be seen that μ is given by the transfer function of the lead-lag network, equn. (19), Part 3, multiplied by K, i.e.

$$\mu = \frac{KpT_2/b}{1 + p(T_1 + T_2/b) + p^2T_1T_2}, \quad (1)$$

where

$$T_1 = C_1(R_1 + R_2),$$

 $T_2 = C_2 R_2 R_1 / (R_1 + R_2).$

With the loop closed, therefore, since $\beta = 1$,

$$\frac{V_{out}}{V_2} = \frac{\mu}{1-\mu} = \frac{KpT_2/b}{1+p\{T_1+(1-K)T_2/b\}+p^2T_1T_2}.$$
(2)

Now the argument used in Part 4 for

deriving equn. (39) from equn. (38) gives, if proper note is taken of the change of suffixes as between Fig. 11(b), Part 4, and the present Fig. 1,

$$\frac{V_2}{V_1} = \frac{1}{pC_2R_2} = \frac{b}{pT_2}.$$
 (3)

Hence

$$\frac{V_{out}}{V_1} = \frac{K}{1 + pT/q + p^2T^2},$$
 (4)

where

$$T^2 = T_1 T_2 \tag{5}$$

and

$$\frac{1}{q} = \left(\frac{T_1}{T_2}\right)^{\frac{1}{2}} + \frac{1-K}{b} \left(\frac{T_2}{T_1}\right)^{\frac{1}{2}}.$$
 (6)

Alternative analysis as a seriesfeedback system

Sallen-and-Key The circuits are commonly used with K = I (nominally), obtained from an amplifier controlled by 100% series negative feedback. Such an amplifier is most simply represented by the cathode follower, as explained in Part I and as used again in the present Part in Fig. 4. Figs. 2(a) and (b) are a reminder of the identity between a cathode follower and a high-gain signinverting amplifier with 100% series feedback, and show that essentially the only difference is in the practical matter of where the circuits are earthed. To clarify the identity the output-current circuit in each case is completed by including R_L and by drawing a short circuit through the h.t. batteries X and any other bias supplies, since it must be assumed that they show negligible impedance to signal frequencies. Ordinarily, of course, the cathode follower, Fig. 2(b), is drawn with the "earthed", or common, line at the bottom. Since

$$K = A/(A+1) \tag{7}$$

 $K \to \mathbf{I}$ only as $A \to \infty$.

By using the enhanced emitter follower, Fig. 3(a), values of A of several

* Royal Radar Establishment.

hundreds are readily obtained; and it may sometimes be useful to extend this type of connection to triples, etc. If an operational amplifier is to be used, one with differential input and which can take 100% feedback is needed, Fig. 3(b). An operational amplifier should give the low voltage drift obtainable from a longtailed-pair input stage, which would be useful in an l.p. filter required to pass zero-frequency (d.c.) signals.

In the ideal case, K = I, equ. (6) reduces to $q = (T_2/T_1)^{\frac{1}{2}}$, and hence $T_2 = qT$ and $T_1 = T/q$. For finite A, substitution from equ. (7) into equ. (6) gives

$$\frac{1}{q} = \left(\frac{T_1}{T_2}\right)^{\frac{1}{2}} + \frac{1}{b(A+1)} \left(\frac{T_2}{T_1}\right)^{\frac{1}{2}}.$$
 (8)

This is the same as equn. (39) of Part 5, and is algebraic proof of the identity of

R

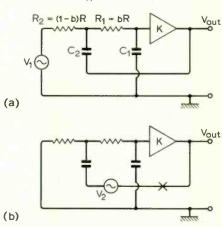
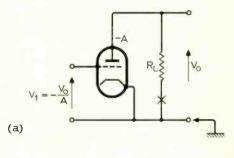


Fig. 1. (a) Sallen-and-Key low-pass filter; (b) the same with V_1 short-circuited and a new source V_2 introduced to facilitate analysis as a lead-lag loop with positive gain.



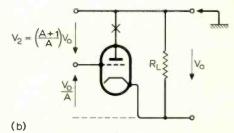


Fig. 2. Example of an active device—a value: (a) in common-cathode connection, i.e. as a high-gain amplifier, $V_0/V_1 = -A$; (b) in common-anode (cathode-follower) connection, i.e. with 100% series negative feedback, $V_0/V_2 = A/(A + 1)$.

the Sallen-and-Key circuit and the lagand-integrator loop with series feedback (Fig. 15 of Part 5), which was mentioned in Part 1. The identity may also be demonstrated graphically as shown in Fig. 4. This is an application of the identity shown in Fig. 2. The only difference between the two circuits is that in (a) terminal 2 of the output is shown earthed, and in (b) terminal I. But since in neither-in so far as the diagrams tell the whole truth about the circuits-does any current flow in the earth lead, the change makes no essential difference and may be regarded as only a device for marking the node which is to be taken as the reference point of potential.

The triode valve in Fig. 4, as elsewhere, is intended as a universal symbol for a three-terminal amplifier. When more complex amplifiers are used, the identity may not be seen so clearly. A multistage amplifier used as a voltage follower will be wired up somewhat differently from when it is used as a sign-inverting amplifier, high-gain because of the practical requirement in each case for operation from an earthed power supply. Similarly if an operational amplifier with differential input is used, the internal workings are somewhat different in the two connections. But as all are close approximations to an ideal three-terminal amplifier the essential identity remains. The two separate drawings of Fig. 4 are, moreover, not really needed. The change of earth point can be made by the disconnection marked X and the reconnection marked with an arrow head.

Compensation for finite internal gain

If A is finite and positive, application of 100% feedback gives K < I, since K = A/(A + I). The theoretically best way of making $K \rightarrow I$ very closely is to make $A \rightarrow \infty$. But in some situations it may be helpful to use an amplifier of moderate internal gain and reduce the feedback, Fig. 5(a), so that

$$K = \frac{A}{A+1} \cdot \frac{r_2 + r_1}{r_2} = 1, \qquad (9)$$

which is obtained when

$$r_2 = r_1/A.$$
 (10)

This artifice, which is easily applied when the amplifier is, for example, an enhanced emitter follower, Fig. 5(b), allows the use of the ideal design values. It is important to remember, however, that K (and consequently q) is just as sensitive to changes in A as before. Caution is needed, therefore, if this method is used to obtain values of q much beyond the reach of the same amplifier without compensation. There is no complete substitute for high internal gain.

In the alternative analysis (or synthesis) (i.e., as a lag and an integrator in a

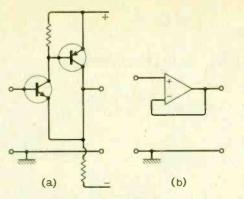


Fig. 3. Possible amplifier configurations for $K \rightarrow I$.

negative-feedback loop) the parallel argument is that finite gain, A, in the integrating amplifier can be compensated by applying positive feedback (feedback fraction I/A) to the amplifier to make its gain apparently infinite, and further that the adjustment is no more critical in the one case than in the other.

In both methods of analysis over compensation produces a regenerative term (negative damping) which subtracts from the positive damping designed into the circuit, and q is higher than intended; but only if the magnitude of the negative term exceeds the positive will the system oscillate, although, of course this is no criterion of satisfactory performance.

It follows also that working the Sallen-and-Key circuit with K > I is equivalent to working the integrating amplifier in the lag-and-integrator circuit with $A > \infty$ (if mathematicians will allow the statement), meaning that the amplifier gain at zero frequency, -A, has gone positive, since A = K(I - K), and that the circuit is working in the region above the diagonal in Fig. 9, Part 5. This further emphasises the regenerative nature of the situation when K > I.

Use of Sallen-and-Key circuit with K > 1

When K = I, $T_2/T_1 = q^2$. Hence when $R_1 = R_2$, $C_2/C_1 = 4q^2$. This may give an inconveniently large value for C_2 . By using K > I a lower value for the ratio T_2/T_1 , and hence of C_2/C_1 , is needed for a given q.

Let $C_1 = C/x$, and $C_2 = xC$. Then $T_1 = CR/x$, $T_2 = xb(1 - b)CR$, and substitution in equal (6) gives

$$\frac{1}{q} = \frac{1}{x} \left(\frac{1}{b(1-b)} \right)^{\frac{1}{2}} + x(1-K) \left(\frac{1-b}{b} \right)^{\frac{1}{2}}, \quad (11)$$

which can be rearranged to give K in terms of x, q, and b. Thus, for example, the circuit may be designed for $C_2 = C_1$, but at the cost of providing components (both Cs and Rs) of sufficient accuracy in initial selection and in long-term stability to meet the increased sensitivity to errors in component values (Fig. 9,

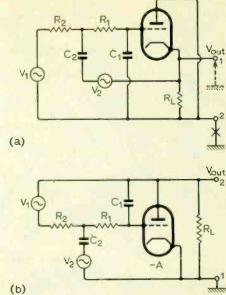


Fig. 4. By changing only the earth point, a Sallen-and-Key l.p. filter is shown to be a lag-and-integrator loop. Also, as the alternative input, V_2 , has one side earthed, the circuit is now suitable for use as a 1st-order band-pass filter.

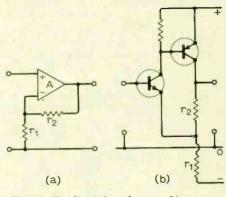


Fig. 5. Feedback less than 100%.

Part 4), which can be interpreted as the result of balancing negative and positive resistance.

Lead-lag network with resistive loading

The network in the feedback path (Fig. 1) is as shown in Part 3, Table 1, diagram (a). As the ratio C_2/C_1 is increased, $k \rightarrow 1$. Hence, when K = 1, the loop gain at ω_c also tends to unity, (i.e., $kK \rightarrow 1$), and $q \rightarrow \infty$ (theoretically without limit) as shown by the diagonal straight line in Fig. 9 of Part 5, q being proportional to $\sqrt{(C_2/C_1)}$.* Up to this limit the active circuit behaves like a passive circuit: no matter what the component values the circuit cannot become unstable (oscillate), and errors in component values are not magnified.

When the network is loaded by resistance R_3 as shown in Fig. 6, $k \ge R_3/(R_3 + R_1)$, and the limiting case

^{*} It is interesting to notice that increasing C_2/C_1 reduces q_0 , the Q factor of the passive network. However, over the useful working region (i.e., to the left of the points of maximum Q) and always when K > I, the increase in k dominates.

is therefore reached when $K = (R_3 + R_1)/R_3$. This may then be considered as a practical maximum value for K, since in general magnification of errors is to be avoided. The presence of R_3 also alters ω_c , and design equations are given in the appendix. This compensation for resistance loading by increasing K will be most useful when the resistances are effectively accurate and stable, and the ratio k is therefore accurately known. If R_3 is the input resistance of the amplifier, it may be subject to considerable uncertainty. It is then desirable that $R_3 \gg R_1$ (and $\gg R_2$), so that if compensation is attempted Kwill be only slightly > 1 (Fig. 7).

High-pass filters

Any low-pass filter can in principle be transformed into a high-pass filter by substituting 1/pT for pT; which means changing a lag into a lead, an integrator into a differentiator, and so on. Operating thus on equn. (4) of Part 5 gives

$$\mu = \frac{pT/q}{1 + pT/q} \cdot pqT \qquad (12)$$

and the schematic shown in Fig. 8. The transformation does not necessarily yield a practical filter however. Fig. 9 is formally the h.-p. counterpart of Fig. 13, Part 5, and if checked by conventional linear circuit analysis gives the expected h.-p. transfer function. As it stands, however, it is unlikely to give a satisfactory performance. At high frequencies it is a shunt feedback system with ratio arms C_1 and C_1' . As the impedances of these fall indefinitely with increasing frequency, and as the response is required to remain level, indefinitely increasing current is called for. An upper limit to these currents can be set by padding out C_1 and C_1' with r and r' inserted at the points X, making $C_1'r' = C_1r$. The presence of these resistors must however to some extent reduce both loop gain and loop phase shift in the region of ω_c , and the circuit is now bener treated as a two-lead loop, time constants C_1r and $C_2 R_2$, with negative gain $(R_1/r \text{ if } A \rightarrow \infty)$, to which the formulae for two lags and negative gain derived in Part 4 can easily be adapted.

The theoretical schematic of Fig. 8 does not show the same difficulty. At high frequencies (well above ω_c) $V_{out} \rightarrow V_{in}$, and the "error" $(V_{in} - V_{out}) \rightarrow 0$. There is therefore no call for indefinitely increasing current through C_1 with increasing ω ; and the same is true for the series-feedback arrangements shown in Fig. 10. Fig. 10(a) shows the functional schematic of a straight-forward circuit with no buffer between the lead and the differentiator; c.f. Fig. 15, Part 5. The design values given are for the ideal case $A \rightarrow \infty$. Reversing the procedure shown in Fig. 4, we redraw the circuit with change of earth point and obtain the Sallen-and-Key high-pass circuit shown in Fig. 10(b). The only serious doubt the

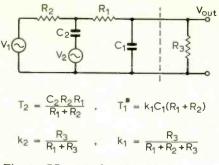


Fig. 6. CR network with resistance loading.

designer should have about these circuits is that in theory the amplifier should have a level response up to infinite frequency. As, however, it can be of the voltage-follower type, and as high internal gain is of importance only in the vicinity of ω_c , it will generally not be difficult to give the amplifier a satisfactory performance up to the highest intended signal frequency. It may, indeed, be thought desirable in these and other active high-pass systems, especially when the internal gain is high, to define the final high-frequency cutoff (with added components) rather than leave it to the chance values of stray capacitances. For finite values of A

$$\frac{1}{q} = \left(\frac{T_2'}{T_1'}\right)^{\frac{1}{2}} + \frac{1}{b(A+1)} \left(\frac{T_1'}{T_2'}\right)^{\frac{1}{2}} \quad (13)$$

The notation is in conformity with Part 3, Fig. 8 and equns. (28) to (31), and the primes serve to draw attention to the inversion of the positions of the suffixes compared with the low-pass case, equn. (8). For K > 1 the appropriate substitutions can be made in the low-pass results.

Input impedance

For low q (say \leq I) the input impedance of a Sallen-and-Key filter is not very different from that of the network when passive. At higher values of q, because T_2/T_1 or T_1'/T_2' become \geq I, at the resonant frequency the voltage across the element behind the input terminal is equal to q times the input voltage approximately, and it is necessary to take account of the relatively heavy current that will flow if the filter itself and the preceding circuit are to operate satisfactorily.

Notch factors

A notch filter with a symmetrical amplitude vs. frequency response may be used to reject a particular frequency, or be combined with others to form a broader band-stop filter. One with an asymmetrical response may be used as a section of a higher-order filter (e.g., Fig. I of Part I) to give a sharper transition from pass band to stop band. In either case the notch is associated with a quadratic factor with a numerator zero.

Passive CR notch networks, with and without buffer amplifiers, have been

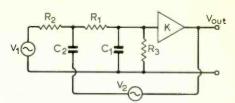


Fig. 7. Resistance-loaded network in active filter.

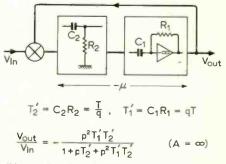


Fig. 8. Lead-and-differentiator loop as h.p. filter.

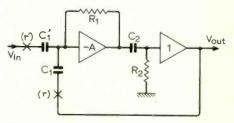


Fig. 9. Practical difficulty in high-pass circuit obtained by direct transformation of low-pass circuit (Part 5, Fig. 13).

described in Part 3. For such networks $q > \frac{1}{2}$; and Fig. 11(a) shows an example which gives a symmetrical notch,

$$\frac{V_{out}}{V_{in}} = \frac{1 + p^2 T_1 T_2}{1 + p(T_1 + T_2/b) + p^2 T_1 T_2},$$
(14)

with a zero at $\omega_{\infty} = 1/(T_1T_2)^{\frac{1}{2}}$, if the necessary equal-time-constant condition is met, $T_3 = T_2$; i.e.,

$$C_3 R_3 = C_2 R_1 R_2 / (R_1 + R_2) \qquad (15)$$

For $q > \frac{1}{2}$ the circuit can be made active (i.e., feedback can be applied) as shown in Fig. 11(b). The part of the circuit above and to the right of the dotted line through X₁ is the standard Sallen-and-Key l.p. circuit (Fig. 1), if the assumption is made that the output impedance of the buffer amplifier (I) is negligible; while the circuit to the left of the dotted line through X₂ is the circuit of Fig. 11(a), unaltered if the assumption is made that the output impedance of the amplifier (K) is also effectively zero. There must therefore be zero transmission at the same frequency as for equn. (14); while signals once injected into the upper part of the circuit, whether through R_2 or through C_1 , are subjected to the q of this active part of the circuit. The complete transfer function is, therefore,

$$\frac{V_{out}}{V_{in}} = \frac{1 + p^2 T^2}{1 + pT/q + p^2 T^2} \quad (16)$$

where $T^2 = T_1T_2$, and q is given by equn. (6). Preferably the amplifier K is a high-gain amplifier with 100% feedback. Then, as before, K = A/(A + 1), and q is given by equn. (8).

For an asymmetrical notch, low-pass type, the numerator becomes $(1 + a'p^2T^2)$, where a' < 1, and

$$\omega_{\infty} = 1/\sqrt{(a'T^2)} = \omega_0/\sqrt{a'}$$
 (17)

 (ω_{∞}) is the frequency of the notch, ω_0 the undamped natural frequency of the system.) The required attenuation in the high-pass path is easily added by connecting the buffer amplifier I to a tap on R_3 ; i.e., the network in box B_1 is replaced by the network shown in Fig. 11(d), which has the transfer $a'pC_3R_3/(1 + pC_3R_3).$ function $T_3 = C_3 R_3$ must of course still $= T_2$, equn. (15). For an asymmetrical notch of high-pass type, attenuation can be introduced into the low-pass path as in Fig. 19 of Part 3, so that, as for the passive network,

$$\omega_{\infty} = \sqrt{a/T} = \omega_0 \sqrt{a} \quad (18)$$

Fig. 11(b) is the standard Sallen-and-Key l.p. filter with an added h.p. path. A notch can just as easily be obtained by taking a standard Sallen-and-Key h.p. filter and adding a l.p. path, in other words by starting from Fig. 15 of Part 3, and turning the lower tee into an active filter. The result is shown in Fig. 11(c). As (theoretically) the gains at zero frequency and at infinite frequency are equal, the response is a symmetrical notch. Fig. 11(e) shows attenuation added into box B₂ of Fig. II(c) (i.e., into the low-pass path) to give asymmetrical notch response, high-pass type. This arrangement may be slightly preferable to that described in the previous paragraph, as high-frequency signals need to pass through only one amplifier.

It is not essential to have a buffer amplifier in the added parallel path. The networks of Fig. 22 of Part 3 can be turned into active filters (e.g., Fig. 12) and the transfer functions are easily derived by making use of those for the passive networks. It is found, however, that for K = I, $q = b'(T_2/T_1)^2$ or $q = b(T_1/T_2)^4$. This results in a greater spread of component values, since b' and b are < I (often $\frac{1}{2}$). Also q_{max} is smaller for a given internal gain A when K = I nominally. It seems likely, therefore, that the circuits with the buffer amplifiers will usually be preferred.

In the filters with a buffer amplifier, q is a function of the active part of the circuit only, and, as in the simple l.p. and h.p. filters, depends on the ratio of two time constants $(T_1 \text{ and } T_2)$ and on the amplifier gain K. T_3 is isolated from the active part of the circuit, and so errors in T_3 do not affect q, although they do affect the accuracy of the required match $(T_3 = T_2)$ and hence the depth of the notch. If the gain of the buffer amplifier (1) is appreciably > or < 1, ω_{∞} is moved accordingly, equns. (17)

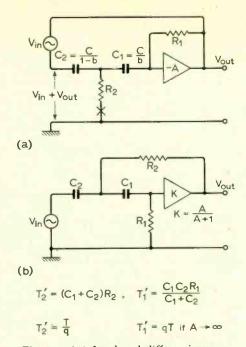


Fig. 10. (a) Lead-and-differentiator loop with series feedback; (b) the same converted by change of earth point to Sallen-and-Key h.p. filter.

and (18), but its gain affects neither q nor the depth of the notch (as long as its output impedance is effectively zero). So in this sense its internal gain is not a critical factor. And in the notch filters without a buffer amplifier, although T_3 cannot vary independently, still q does not depend critically on the balance of components or of time constants, at least for $K \leq I$. This contrasts with the behaviour of some rather similar-looking circuits based on the CR parallel-tee network, which can give higher values of q for a given value of the internal gain, A, when K = I nominally, and which will be the subject of a later article.

Simple bandpass (tuned-circuit) response

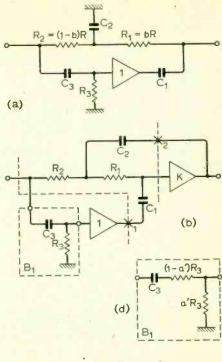
As shown at the beginning of this article, Fig. 1(b) and equn. (1), the standard Sallen-and-Key l.p. filter can be arranged to give tuned-circuit (1st-order bandpass) response by injecting the signal voltage in series with C_2 ; and, as shown in Fig. 4, for the case K = A/(A + I), (i.e., $K \leq I$), by moving the earth point a more convenient arrangement with one side of V_2 earthed is obtained. Similarly the h.p. filter, Fig. 10, is converted to band-pass by injecting the input at the point X. The strange appearance of the circuits is partly remedied by a change of layout as shown in Figs. 13(a) and 14(a), which show the circuits as having feedback networks of familiar form (e.g., Fig. 3, Part 1), only the input connections being unusual.

Because of the limitation

$$q_{max} = \frac{1}{2}\sqrt{b(A+1)}$$
 or $\frac{1}{2}\sqrt{b'(A+1)}$

the circuits are likely to be of limited application; for band-pass filters usually require higher Q factors than low-pass or





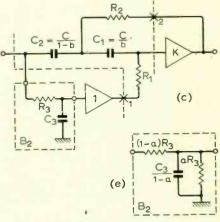


Fig. 11. Modifications for obtaining a notch (or zero): (a) Effectively passive symmetrical notch filter with buffer amplifier in h.p. path; (b) the same made active; (c) similar circuit with buffer in the l.p. path; (d) a method of putting attenuation in the h.p. path of (b); (e) in the l.p. path of (c).

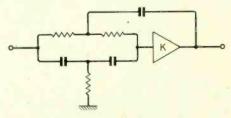


Fig. 12. Notch filter without buffer.

high-pass. For values of q well below q_{max} , however, the circuits are interesting in being fully "designable" while using the minimum possible number of components, one amplifier, two capacitors, and two resistors. This very economy, however, makes the circuits unaccommodating; e.g. the gain at resonance which for K = I is equal to q^2/b or q^2/b' (i.e., $2q^2$ when b and $b' = \frac{1}{2}$) cannot be varied independently of q;

Wireless World, January 1970

and to overcome this inflexibility additional components must be added.

An *ab initio* analysis of the circuits, for $A = \infty$, K = I, could be made as follows: (a) write down the current flow caused by V_2 , assuming both the input of the amplifier (the virtual earth) and the output are shorted to ground; (b) write down the current flow caused by V_{out} , assuming that both the virtual earth and V_2 are shorted to ground; (c) set the sum of the currents converging on the virtual earth to zero.

The useful part of the current flow (a), i.e. the equivalent exciting current, is the current through R_1 (in Fig. 13) and through C_1 (in Fig. 14), i.e.

$$I_{equiv} = \frac{V_2 p T_2}{R_1 (1 + p T_2)}$$
(19)

and
$$I_{equiv} = \frac{V_2 p_1 C_1}{1 + p T_2}$$
 (20)

where T_2 is the time constant given by C_2 in combination with R_1 and R_2 in parallel (Fig. 13), and T_2' the time constant given by R_2 in combination with C_1 and C_2 in parallel (Fig. 14).

These equations, as could be forseen, represent the current a voltage source would drive through a series CR branch; and so the same response will be obtained if the source V_2 is replaced by the source V_2'' feeding in through a branch C_3 , R_3 , of time constant as specified, Figs. 13(c), 14(c). For the magnitude (gain) to be the same, R_3 should equal R_1 , or C_3 should equal C_1 , for the two cases respectively. But the advantage of having the added branch is that now the gain can be varied independently by varying the impedance of the branch while keeping the product C_3R_3 constant.

Alternatively the response to an input V_2' may be calculated. Thus, for Fig. 13,

$$\frac{V_{out}}{V_2} = -\frac{R_1 + R_2}{R_3} \cdot \frac{1 + qpT}{1 + pT/q + p^2T^2};$$
(21)

whence by equating currents the response to $V_2^{"}$ can be obtained,

$$\frac{V_{out}}{V_2''} = -\frac{R_1 + R_2}{R_3} \cdot \frac{qpT}{1 + pT/q + p^2T^2}$$
(22)

if $C_3R_3 = qT = T_2 = C_2R_1R_2/(R_1 + R_2)$.

By the same type of argument it is easily shown, Fig. 13(d), that low-pass response may be obtained by feeding in a signal $V_2^{"}$ through a low-pass (simplelag) tee network, again of the same time constant, $T_2 = qT$.

From the other circuit similar derivations can be made as indicated in Fig. 14. Here it is found that

$$\frac{V_{out}}{V_{2}'} = -\frac{R_1}{R_3} \cdot \frac{1 + pT/q}{1 + pT/q + p^2T^2}$$
(23)

and consequently that C_3R_3 should now equal $T/q = T_2' = (C_1 + C_2)R_2$.

Although all these derived circuits are extravagent in the number of components used, they can be a convenient practical choice. The extra components cause some reduction in q_{max} , but usually the effect is slight. Component values for C_3R_3 are not critical, since the input branch is effectively isolated from the resonant feedback loop by the virtual earth and has almost no effect on q.

Appendix

The transfer function giving V_{out}/V_2 for Fig. 6 is readily obtained from equn. (1) by substituting the impedance of the parallel combination of R_3 and C_1 , i.e. $R_3/(1 + pC_1R_3)$, for $1/pC_1$; and by making the following convenient substitutions:

$$k_1 = R_3/(R_1 + R_2 + R_3),$$

$$k_2 = R_3/(R_1 + R_3),$$

$$T_1^* = k_1T_1,$$

i.e., T_1^* is the CR product formed from C_1 and $(R_1 + R_2)$ in parallel with R_3 . [Note: $k_1/k_2 = k_1(1 - b) + b$]. Thus it is found that

$$\frac{V_{out}}{V_2} = \frac{\frac{k_1 p T_2 / b}{1 + p \left(T_1^* + \frac{T_2 k_1}{b k_2}\right) + p^2 T_1^* T_2}}{(24)}$$

From this, proceeding as before, equns. (1) to (4), the l.p. transfer functon for the active circuit, Fig. 7, is obtained as

$$\frac{V_{out}}{V_1} = \frac{k_1 K}{1 + p T/q + p^2 T^2},$$
 (25)

where

$$T^2 = T_1^* T_2^{}, (26)$$

$$\frac{1/q}{(k_1/k_2)(1-k_2K)(T_2/T_1^*)^{\frac{1}{2}}}.$$
(27)

In the special case $K = I/k_2$, $I/q = (T_1^*/T_2)^{\frac{1}{2}}$ and $T_1^* = T/q$, $T_2 = qT$, which have the ideal form of the corresponding equations for a simple *LCR* passive prototype. Thus the sensitivity to errors in capacitor values is the same as that of the circuit with an unloaded network, although, since K > I, there is additional sensitivity to errors in the values of the resistors that determine K and k_2 .

Corrections to Parts 3 and 4

In Part 3, October issue, in the caption to Fig. 16 CT_2 was printed instead of CR_2 .

In Part 4, November issue, the following have been noticed. In Fig. 2, in the box representing the passive network, only the denominator of the transfer function appears. This should read I/(the expression printed). In Fig. 5(a), I + has been omitted from the

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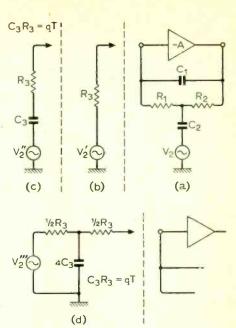


Fig. 13. Derivation of band-pass filter and an alternative form of low-pass filter from the standard l.p. filter (Fig. 4).

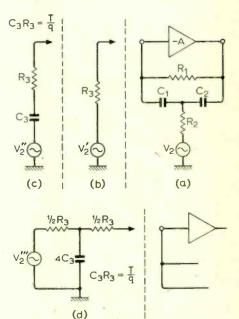


Fig. 14. Similar derivations from the standard h.p. filter (Fig. 10).

denominator of the transfer function in the box representing the passive network; and in the last full column (p. 525) references to Figs. 10(a) and 10(b) should read 11(a) and 11(b) respectively.

Personalities

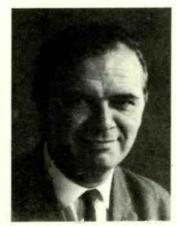
Professor C. W. Oatley, O.B.E., F.R.S., professor of electrical engineering in the University of Cambridge, has been awarded one of the Royal Society's three Royal Medals for 1969/70 "for his distinguished work in the wartime development of radar and latterly for the design and development of a highly successful scanning electron microscope." Professor Oatley has occupied the chair of electrical engineering at Cam-bridge since 1960. He was in charge of basic work on radar transmitters and receivers at the Government Radar Research & Development Establishment during the war and from 1945 until receiving his professorship was lecturer in electrical engineering at the University.

The University of Edinburgh has appointed P. L. Kirby, D.Sc., F.Inst.P., who is research director of Welwyn Electric Ltd, of Bedlington, Northumberland, as its second visiting industrial professor in the newly established Microelectronics Liaison Unit within the School of Engineering Science. Dr. Kirby graduated from Durham University during the war and after two years working on radar systems at T.R.E., Malvern, returned to the North East where he worked on the physical properties of glasses. During this period in industry he took the further degrees of M.Sc., and D.Sc. from Durham and then in 1956 moved to Welwyn Electric Ltd. Professor Kirby has maintained a personal interest in the measurement and interpretation of noise and non-linearity effects in resistive materials.

Peggy Lilian Hodges, head of Guided Weapon Simulation and System Analysis at the Stanmore Laboratories of GEC-AEI (Electronics), has been elected to the Fellowship of the Royal Acronautical Society in recognition of the contribution she has made to avionic and guided weapon technology. Miss Hodges was born in 1921 and educated at Westcliff High School, Essex, and Girton College, Cambridge. She joined GEC-AEI (Electronics) in 1950, and her work at Stanmore has been centred largely on the performance of guided weapons. She has worked on many projects, notably Seaslug and Sea Dart, with a particular interest in weapon simulation techniques. Miss Hodges is a senior vice-president of the Women's Engineering Society.

Charles Kao, B.Sc. (Eng.), Ph.D., M.I.E.E., has been appointed an honorary senior research fellow at Queen Mary College, University of London. This is the second such appointment from Standard Telecommunication Laboratories in recent months with a view to bringing industrial experience to University affairs. From time to time Dr. Kao will lecture on his specialist subjects including topics in optical communications, coherent wave optics, and electromagnetic problems. Dr. Kao, who is 36, has been with S.T.L. since 1961, where latterly he has been mainly concerned with problems associated with the transmission of coherent light down optical waveguides for future telecommunication systems.

Alan Hall has joined Oxley Developments Company, of Ulverston, Lancs, as promotional sales manager. He was until recently in the electronic component division of Johnson Matthey, prior to which he was with Muirhead. Mr. Hall operates an amateur radio station with the call G3UWA.



Alan Hall

John L. Carroll, who joined Data Recognition Ltd in 1966 as general manager and has been responsible for the development of their current range of optical document readers, has been appointed technical director. Prior to joining Data Recognition, Mr. Carroll was with English Electric Computers where he was responsible for the development of document handling peripheral equipment; and before that he worked for Solartron on the development of character recognition equipment.

G. Ross Watson appointed by the Video Systems Division of Bell & Howell as international marketing manager, was until recently sales engineer responsible for marketing television camera tubes with the English Electric Valve Company. For three years from 1958 Mr. Watson, who is 44, was manager of a mobile television unit frequently used to demonstrate the value of closed-circuit colour TV at surgical operations.

Peter Smitham who joined ITT Electronic Services, Harlow, Essex, in July 1967 has been appointed manager. He was previously materials manager. He studied at University College Swansea and spent a postgraduate year at Salford University.

William J. Charnley, appointed deputy controller of guided weapons in the Ministry of Technology, was educated at Oulton High School, Liverpool, and at Liverpool University where he obtained a first class honours degree in engineering. Mr. Charnley, who is 47, joined the Civil Service in 1943 at the Royal Aircraft Establishment, Farnborough. He was appointed superintendent of the Blind Landing Experimental Unit in 1955. Six years ago he became head of the Instruments and Electrical Engineering Department at R.A.E. and two years later was appointed head of the Weapons Department. Since 1968 he has been head of the Establishment's Research Planning Division where he is succeeded by Harold G. Robinson, O.B.E., who has been head of the Avionics Department since 1965. Mr. Robinson, who is 45, was educated at H.M. Dockyard School, Portsmouth, where he was awarded a Whitworth Scholarship to Imperial College, London University. He. obtained a 1st class honours degree in electrical engineering, and after joining the Civil Service at the R.A.E. in 1948 continued his post-graduate studies during 1951-52 at the Californian Institute of Technology. In 1955 he took charge of the Black Knight research rocket project. From 1960 until 1965 he was in charge of the satellite launcher division at Farnborough.

Tudor Jones, M.I.E.E., aged 42, has joined Cambion Electronic Products Ltd, manufacturers of electronic components, as sales manager. He joins Cambion from the English Electric Co., Stafford, where he was manager of the Production Systems Department.

Peter L. Mothersole, F.I.E.R.E., M.I.E.E., has joined Pye T.V.T. Ltd, Weybridge, as engineering manager of the Audio & Vision Division. Mr. Mothersole, who is 40, has been with the Mullard



Peter Mothersole

Research Laboratories since 1953, where he was for some time leader of the television receiver group. During his National Service he was a radar theory instructor at R.A.F. Yatesbury and then spent a year with E. K. Cole Ltd at Malmesbury as a design engineer on airborne radar equipment.

Paul Spring, who joined Grundig (Great Britain) Ltd on its formation in 1952, has been appointed managing director. Mr. Spring has successively been chief engineer, general works manager and, since 1964, technical director.

Geoffrey E. Beck, B.Sc., F.I.E.E., for the past two years chief engineer of Marconi's Electronics Group is appointed technical manager of its Aeronautical Division, based at Basildon, Essex. Mr. Beck, who is 53, graduated at Birmingham University in 1938 and joined the Marconi Research Division, where he worked on the design of naval radar equipment thoughout the war. In 1949, Mr. Beck began his long association with the pioneering work into the development of Doppler navigation equipment, which provides pilots with continuous positional information without the use of ground-based aids. In 1962, Geoffrey Beck and Mervyn Morgan, who were jointly responsible for this work, were awarded the Johnston Memorial Trophy by the Guild of Air Pilots and Air Navigators in recognition of their service to aerial navigation. From 1965 to 67, Mr. Beck was manager of the group responsible for the development of the television guidance system for the Martel guided missile. He is vice-president of the Institute of Navigation.

4

Progress in Tape-recording Techniques

by Sidney Feldman

Exhibits at the Audio Engineering Society's 37th Convention, held in New York City in October, were predominantly of interest to recording studio engineers. Several 8-track one-inch and 16-track two-inch studio recorders were on operational display. Two machines (Gauss and Magnetic Recording Systems) employed d.c. capstan-drive systems, with the motor-speed controlled precisely by a magnetic tachometer referred to a high stability oscillator. Thus the recorder is not affected by power line frequency variations. The Magnetic Recording machine has a switch-selected speed-range of 32 to 1, permitting operation from $1\frac{7}{8}$ i.p.s. to 60 i.p.s., and also allows any intermediate speed to be obtained using an external variable-frequency source. Fig.1 shows the basic servo-system employed.

In the Gauss recorder, external synchronization is possible for variable-speed operation, with possible pitch changes of $\pm 75\%$. This tape machine utilizes the "focused gap" system of recording, which was marketed, under licence, by Fairchild Recording several years ago, in a series of tape recorders. The bias frequency is approximately 1MHz, and specifications call for a signal-to-noise ratio of 70dB, record input to reproduced output, measured with ASA curve A; peak record level set for 1% distortion on 3M Company 201 tape, at 15 i.p.s.

Of the high-speed tape duplication equipment, Gauss utilizes an endlessloop tape bin, "focused-gap" head, and a bias frequency of 10MHz. Duplication takes place at speeds to 240 i.p.s. Running from a 1200ft master tape at the highest speed, this system can produce 55 copies/ hour /slave, utilizing the "stagger loading" system at the slaves. These copies would be at $1\frac{7}{8}$ i.p.s. The tape would then have to be loaded into the appropriate cassette or cartridge. Console designers are now using, mainly, operational amplifiers in their modules. The modules are completely wired by the manufacturer, saving labour and inter-wiring when a system is built-up. These console "building-blocks" are usually strips 1.5in wide and about 14in long, and they provide functions of equalization, reverberation level control, main-channel level control, input attenuator, and microphone/ line input switching. A typical module will accept microphone level at the input and provide up to +24 dBm output with less than 0.5% t.h.d. from 20Hz to 20kHz.

The large recording consoles in use today with loss-less mixing, are only possible using operational amplifiers. A typical mixing circuit, as shown by Melcor (Fig. 2), provides 114dB of isolation between inputs at 20kHZ, rising to 134dB at 1kHz. Distortion is 0.25% from 20Hz to 20kHz, at full output of + 20dBm. Gain can be adjusted to a maximum of 10dB.

Most tape recorders for the professional market are using transistors for switching functions, and the Quad-Eight Company even have a logic system for track switching on their large console designs. This logic switching can also be interlocked with the tape recorder and the monitoring system, so that operating one button will switch all functions simultaneously.

Distortion analyser

Crown International have developed an i.m. distortion analyser to test, on a production basis, the Crown DC-300 dualchannel amplifier. Typical i.m. distortion, per channel, (60Hz-7kHz, mixed 4:1) is below 0.05% from 0.01 watt to 150 watts r.m.s. into EQ. The analyser permits rapid measurements of i.m. distortion over a wide range of input levels and power ratings. Active Butterworth filters replace conventional hum-sensitive LC filters. The residual distortion in the analyser itself is typically 0.003%. Ganged input and output controls are employed to facilitate production line testing of amplifiers and other equipment.

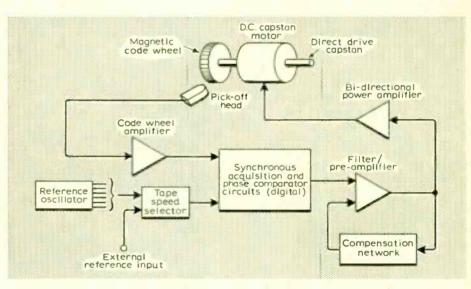


Fig. 1. Speed control system employing magnetic tachometer.

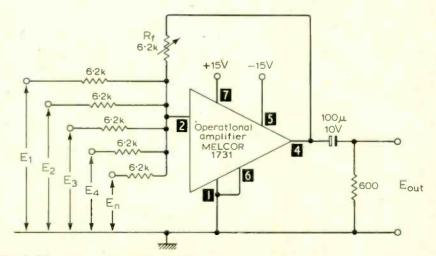


Fig. 2. Mixer using operational amplifier.

World of Amateur Radio

Licences and "pirates"

Of every five new British amateur licences now being issued, rather more than three are for Class B operation on v.h.f. telephony only. The licence statistics to the end of October show that in the previous six months there was an increase of 311 Class B licences to a total of 1841, compared with an increase of 198 Class A licences (permitting h.f. operation and requiring the passing of a Morse test) to a total of 13,373. Taking into account the 180 amateur television licences British amateurs now total nearly 15,400.

That there is still an appreciable number of people who attempt to operate in the amateur bands without the formality of a licence is shown by the fact that, in the first nine months of 1969, the Post Office successfully prosecuted more than 70 persons and warned 50 others for offences involving wireless transmitting apparatus being used contrary to Section 1 of the Wireless Telegraphy Act. The fines imposed on these "pirate" stations ranged up to about $\pounds100$ and the penalties often included forfeiture of the apparatus. The Post Office has indicated that some of the illicit transmissions are regarded as representing a potential hazard to safety of life; they are equally unpopular with licensed amateurs who have sometimes been subjected to deliberate interference and embarrassed by such tricks as the tape recording of genuine amateur "contacts" which are then replayed on transmissions outside the limits of the amateur bands. The Post Office welcomes information which would help in tracing and apprehending the pirates.

Good tropospheric "openings"

During the spell of pronounced tropospheric propagation in mid-October—conditions which brought many complaints of co-channel interference to broadcasting organizations large numbers of amateur v.h.f. contacts were made between stations in the U.K. and many countries of Western Europe. The contacts extended from Sweden and Finland to Austria and Switzerland as well as the almost routine links with France, Holland and West Germany. While most of the contacts were made in the 144-MHz (two-metre) band, the conditions extended also to the 432-MHz (70 cm) band; many of the contacts exceeded 1000 miles. Some amateurs believe that the "tropo" conditions of the 1969 Indian summer were among the most pronounced yet recorded. Peter Blair, G3LTF, of Chelmsford, has raised his total of countries worked on 144 MHz to 28 and on 432 MHz to 19. From January 1st the new voluntary divisions of the 144-MHz band are: 144.0-144.5 telegraphy only; 144.15-144.5 south-west region; 144.5-145.1 south-east region; 145.1-145.5 midlands; 145.5-145.95 north, Scotland and Northern Ireland; and 145.95-146 beacon transmissions.

R.S.G.B. president for 1970

Dr. John Saxton, director of the Science Research Council's Radio and Space Research Station (Slough) and this year's chairman of the I.E.E. Electronics Division, is to be installed as the 36th president of the Radio Society of Great Britain during the course of a social evening at the Bonnington Hotel, London, on Friday, January 16th. Dr. Saxton, although not himself the holder of an amateur licence, has a keen interest in the relationship between meteorology and radio propagation, and for a number of years has attended many amateur v.h.f. functions.

It is clear from the latest Society accounts that the 1970 Council, despite the recruitment in recent years of several thousands of additional members, faces problems of the type which are seriously affecting many national and local societies. In each of the past four yars expenditure has exceeded income: to a total in the four years of some $\pounds 8500$. Fortunately, the Society has substantial reserves but nevertheless it faces acutely the paradox that even with the present record membership, costs are still rising faster than revenue.

In Brief: There has been a good response from British amateurs to a proposal -"Project Trident"-to design and build in the U.K. an amateur radio communications satellite capable of receiving on the 144-MHz band and retransmitting the signals on the 432-MHz band; it is recognized that such a project will take a considerable time to complete ... It is now hoped that the Australian-built amateur satellite "Australis-Oscar 5" will be launched during December or early January on the Thor-Delta rocket used to put a Tiros weather satellite into orbit; Australis is expected to radiate on 144.050MHz and 29.450MHz for a number of weeks . . . The R.S.G.B. 1.8 MHz Affiliated Societies' Contest is due to be held on January 10th and 11th between 18.00 and 22.00 G.M.T. each day . . . This season's 1.8MHz "Transatlantic Tests" will be continued between 05.00 and 07.30 G.M.T. on December 28th, January 11th and February 1st and 15th . A Boy Scouts station on the Caribbean Island of Anguilla, with the callsign VP2EQ, is being operated in the 14-MHz band . . . Indian stations have recently been using the special prefix VU0 instead of VU2 as part of the Ghandi centenary celebrations . . . Mike Matthews, G3JFF, is operating as a maritime mobile station on board the Far East flagship H.M.S. London and is expected to visit many Far East and Pacific areas during the next 15 months. . . The Royal Naval Amateur Radio Society, with the callsign G3BZU, transmits a monthly Morse pro-ficiency test at 19.00 G.M.T. on the first Tuesday of each month on

MHz for speed proficiency tests. PAT HAWKER, G3VA

1.875MHz for practice runs and 3.520

Since the war, amateur radio in Japan has developed greatly. Station of Yoshio Sameshima, JA2CLI, has many 1.8MHz ("Top Band") achievements to its credit; contacts include the U.S., Hawaii, Canada and Australia (photo. courtesy of Stewart Perry W1BB).



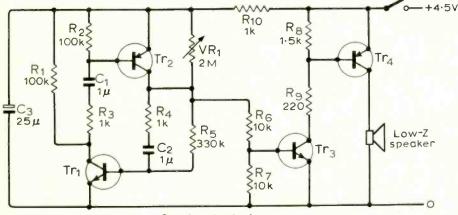
Electronic Metronome

An efficient circuit giving a speed range of 30 to 240 beats per minute

by D. T. Smith

The timing pulses are generated by a complementary-pair form of multivibrator Tr_1 , Tr_2 . Both transistors conduct or are cut off together and the conduction time can be made a very small part of each cycle. The conduction time (about 10ms) is used to generate the "tick" and the off time (0.25 to 2s) the interval

As this form of multivibrator may be unfamiliar to some readers, its operating cycle is described. During the conduction period the base of Tr_1 is forward biased and therefore held near zero volts, while Tr_2 is conducting hard and charging C_2 via R_4 towards +4.5V. The conduction period of Tr_2 is limited by the time C_1 can



Complete circuit of metronome.

between ticks. This form of multivibrator is useful in many applications where short pulses, relative to their separation, are required. Low frequencies can be obtained with relatively low capacitor values, and a good range of frequencies obtained by varying a single resistor.

supply enough base current to drive Tr_2 . This time is set primarily by C_1 and R_3 . When Tr_2 is cut off R_6 and R_7 pull the collector of Tr_2 down to zero so that the charge on C_2 gives a negative voltage on the base of Tr_1 . Tr_1 is thus cut off, and no base current flows. The off time is that taken for VR_1 and R_5 to discharge C_2 and then generate a forward bias at the base of Tr_1 to start conduction. As current starts to flow in Tr_1 it feeds current via R_3 and C_1 to the base of Tr_2 which in turn feeds current via R_4 and C_2 to the base of Tr_1 and so closes a regenerative feedback loop Both transistors are switched hard on, and the cycle is repeated.

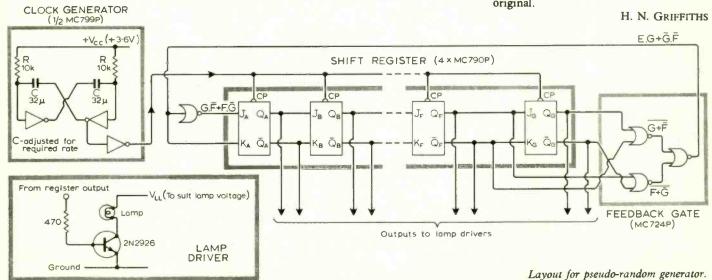
There is no point in having a good quality amplifier to drive the loudspeaker, so a simple two-stage amplifier Tr_3 and Tr_4 is used. This is direct coupled with both transistors cut off when the multivibrator transistors are off, and both conducting hard when the multivibrator transistors are conducting. Any medium- or high-gain silicon planar transistors can be used for Tr1, Tr2 and Tr₃. A medium or high power germanium p-n-p transistor capable of switching about 1A-AC128, OC83, OC25 etc.can be used for Tr_4 .

There are no special precautions necessary in wiring the circuit, and Veroboard is convenient to use. C_2 should be a paper or polyester type and not electrolytic, as variations of capacitance with age or leakage would upset the calibration. The circuit with loudspeaker and battery are fitted in a suitable box. A low value variable resistor, say 25 3 may be included in series with the speaker to act as a volume control.

A Digital Christmas Tree

Using a large number of many coloured lamps switched in pseudo-random sequence, an elegant and fascinating twinkling effect is produced which can be exploited by arranging the lamps around a Christmas tree. In the prototype a 7-stage shift register is used to generate the 'M' sequence, an exclusive OR gate being used to provide feedback to the input from stages 6 and 7. The length of the sequence is 127 clock periods (27-1). It is arranged that fourteen lamps are controlled by transistor switches from the shift register stages (7 from the Q and 7 from the \overline{Q} outputs). Transistors can, of course, replace the micrologic elements in the sequence generator. Small pea lamp bulbs (model railway type) are used in the original.

H. N. GRIFFITHS



D.C. Bias in Push-Pull Power Amplifiers

A feedback amplifier controls the working point of a directly coupled driver and output pair

by R. A. Smith

A bias circuit, conventional to many amplifiers, and consisting of d.c. feedback from the collector to the base of the driver transistor, is shown in Fig. 1. A typical practical arrangement is given in Fig. 2. However, the circuit is very sensitive to changes in the current gain of the driver transistor, and hence R_f must be adjusted for the particular driver transistor used.

Since the prices of suitable transistors are now very low, extra low-current transistors in a circuit present no problem. Hence a transistor can be used whose sole purpose is to stabilize the d.c. level of the output. The requirements of this transistor are:—

(i) High output impedance, so as not to shunt the a.c. signal at the base of the driver transistor.

(ii) A positive current gain to ensure that the d.c. feedback is negative. (The driver transistor changes the sign of the signal).

Its input impedance is not important since it is to be driven through a filter from the low-impedance output of the voltage amplifier, which is capable of supplying large currents; i.e., its loading effect on the amplifier output will be negligible even if the extra transistor has a low input impedance.

In order to satisfy (i), the collector of the stabilizer transistor must be connected to the driver's base; in order to satisfy (ii) a common base circuit must be used. For such a transistor to conduct at all, it must be of opposite polarity to the driver transistor. Also, the a.c. level of the output must be filtered from the d.c. to prevent excessive a.c. feedback. Using a simple RC circuit for this, we obtain the arrangement shown in Fig. 3.

 R_2 must be chosen so that the potential across it is small compared with the supply voltage. The current passing through this resistor is approximately the base current of the driver transistor, i.e. $1/\beta$ multiplied by driver transistor collector current which is approximately $V/2R_L\beta$.

Thus $V \gg V_{R_2} = iR_2 = R_2 V/2R_L\beta$

 $2R_L\beta/R_2 \gg 1$ and $R_2 \ll 2R_L\beta_{min}$ in the worst case.

With R_L as $2k\Omega$, $\beta_{min} = 20$ and a factor 20 for " \ll ", this gives $R_2 = 4k\Omega$.

The R_2C time constant must be long compared with the lowest frequencies being used.

Fig. 4 shows a circuit used as an a.f. power amplifier. The values given are for a 3Ω speaker and 25-V power supply.

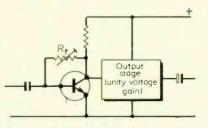


Fig. 1. Arrangement where driver stage Tr_1 controls d.c. level of output pair.

High output impedance amplifier

With high output impedance amplifiers we are trying to drive a current in a load irrespective of the potential across it. In an inductive load, for example, the potential may vary considerably depending on the waveform of the input. For the output to be at high impedance, the drive should be from the collectors of the

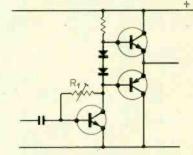


Fig. 2. Typical low-to-medium power output stage.

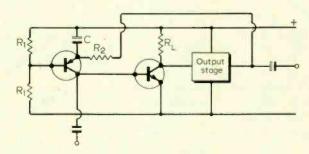
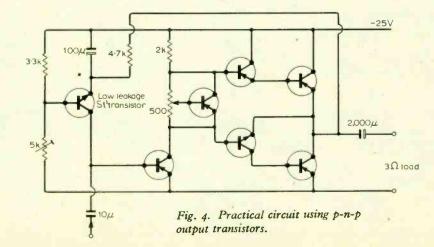


Fig. 3. Use of common-base stage to control d.c. level of amplifier output.



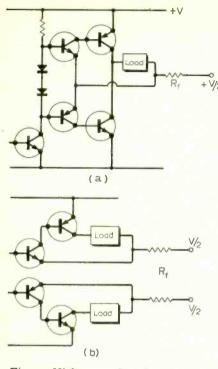


Fig. 5. High output impedance power stage. (a) Actual configuration, (b) Convenient bisection for analysis.

Fig. 6. Application of bias transistor to circuit of Fig. 5.

transistors, since the currents in these are the least sensitive to variations of potentials across them.

To apply feedback, the current in the load is passed through a resistor in series with the load, as shown in Figs. 5(a) and 5(b), and hence converted into a voltage which is compared with the input voltage to the amplifier at the emitterbase junctions of one of the driver transistors, depending upon which half of the circuit is conducting.

As will be seen from Fig. 5(b) each part, upper and lower, is an emitter follower across R_f ; however, the largest part of the current in R_f comes from the collector of the output transistors (high impedance). There is also a small current in R_f which flows in the emitter of whichever driver transistor is conducting and this current is approximately the same as the collector current, i.e. $1/\beta$ times the current through the load in series with R_f .

 R_f should be chosen so that the maximum current in the load (and hence also in R_f) produces a voltage across R_f of, say, a tenth of the power supply voltage. In this case, there is a maximum potential swing of approximately ± 0.4 of the supply voltage across the load, and the back e.m.f. of an inductive load must not exceed this value if the amplifier is to work in its linear region.

The bias circuit described above can

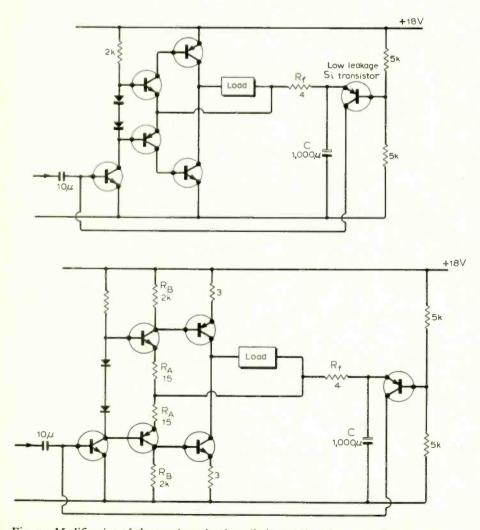


Fig. 7. Modification of the previous circuit to limit standing current in the output stage.

be incorporated in the circuit as shown in Fig. 6.

In a practical circuit, the standing current in the output transistors can become intolerably large if bias resistors are not included as shown in Fig. 7. Resistors R_A reduce the quiescent current of the driver transistors and resistors R_B bleed much of the remaining driver transistors' current to the supply rails rather than through the bases of the power transistors. Values shown are for an amplifier driving 0.25A into an inductive load; it was for driving the scan coils of a magnetically deflected c.r.o.

More Announcements (see also p.19)

Hewlett-Packard Ltd have installed a £20,000 patient monitoring system at Walsgrave General Hospital, Coventry. The system includes a cardiac catheterization unit, bedside monitors for coronary care, defibrillator trolleys and two nurse's central stations.

Pye Telecommunications Ltd, has received contracts worth $\pounds 575,000$ for radio telephone equipment to be supplied to Government departments. The orders include u.h.f. fixed and mobile f.m. equipment for airports, v.h.f. motorcycle and personal transmitter /receivers, and v.h.f., h.f. and m.f. marine transmitter /receivers.

STC Mobile Radio Telephones Ltd has been awarded a contract valued at approximately £700,000 by the Home Office for the supply of several thousand mobile a.m. radio telephones for police use.

Plessey Electronics Group has received an order from the North of Scotland Hydro-Electric Board for the supply of radio relay equipment. Operating in the 1500MHz band, the equipment will link the Board's head office in Edinburgh with its main control room at Port-na-Craig, Pitlochry, and communications centres at Burghmuir, Perth and Tealing near Dundee.

A telecommunications project. A contract worth \pounds 1.4M has been awarded to S.T.C. by the Ministry of Posts, Telegraphs and Telephones of Kuwait for the provision of a number of broadband microwave and coaxial cable links from Basra, Iraq, into the telecommunications centre at Kuwait City.

Microwave Associates Ltd, have announced an agreement whereby they will provide a complete marketing service for the products of Microwave Semiconductor Corporation, Somerset, New Jersey, U.S.A.

Electron microscopes worth more than $\$ \frac{1}{2}M$ have been ordered from AEI Scientific Apparatus Ltd for the United States and Canada. These orders have been placed by Picker Nuclear, the company's associates in the United States.

1970 U.K. Conferences and Exhibitions

Further details are obtainable from the addresses in parentheses

LONDON

- Jan. 19–23 Bloomsbury Centre Hotel American Data Communications Equipment (U.S. Embassy, Grosvenor Sq., London W1A 1AE)
- Mar. 2-5 Physics Exhibition Alexandra Palace
- (I.P.P.S., 47 Belgrave Sq., London S.W.1) Mar. 10–12 Camden Town Hall Sound '70 International (Association of Public Address Engineers, 394 Northolt Rd., South Harrow, Middx.) Mar. 17–19 Savoy Place
- Mar. 17–19 Savoy Pl Electrical Methods of Machining, Forming and Goating (I.E.E., Savoy Pl., London W.C.2)
- Apr. 8–15 Earls Court Electrex '70 (Electrical Engineers A.S.E.E. Exhibition, Museum St., London W.C.1)
- Apr. 13-16 University College Atomic and Molecular Physics (I.P.P.S., 47 Belgrave Sq., London S.W.1)
- Apr. 23–26 Skyway Hotel High Fidelity Exhibition (Federation of British Audio, 49 Russell Sq., London W.C.1)
- Apr. 28 & 29 Royal Garden Hotel Microelectronics Conference (Business Conferences & Exhibitions,
- Mercury House, Waterloo Rd., London S.E.1) May 4–7 Royal Festival Hall London Engineering Congress (LECO 70) (Council of Engineering Institutions, 2 Little Smith St., London S.W.1)
- May 5-15 Earls Court Mechanical Handling Exhibition (Iliffe Exhibitions, Dorset House, Stamford St., London S.E.1)
- May 11–13 Middlesex Hosp. Med. School Television Measuring Techniques (I.E.R.E., 8-9 Bedford Sq., London W.C.1)
- May 11–16 Olympia Instruments, Electronics & Automation Show (Industrial Exhibitions, 9 Argyll St.,
- London W.1) May 19-21 Savoy Place Signal Processing Methods for
- Radio Telephony (I.E.E., Savoy Pl., London W.C.2) June 9-11 Electrical Interference in Instrumentation
- (1.E.E., Savoy Pl., London W.C.2) July 13–17 Olympia
- Ships' Gear International Show (Brintex Exhibitions, 3 Clements Inn, London W.C.2). Sept. 7-11 Grosvenor House
- International Broadcasting Convention (International Broadcasting Convention, Savoy Pl., London W.C.2) Sept. 15–18 Olympia
- Bio-Medical Engineering Exhibition (U.T.P. Exhibitions, 36-37 Furnival St., London E.C.4) Sept. 15-18 Savoy Place
- Electrical Discharges in Gas (I.E.E., Savoy Pl., London W.C.2)

- Sept. 29-Oct. 2 Savoy Place Trunk Telecommunications by Guided Waves
- (I.E.E., Savoy Pl., London W.C.2) Oct. 15-21 Audio Festival & Fair (International Audio Festivals and Fairs, 42 Manchester St., London W.1)
- BANGOR July 6–10 University College
- Microwave Spectroscopy (1.P.P.S., 47 Belgrave Sq., London S.W.1)

BIRMINGHAM

Apr. 14–16 Automatic Test Systems (I.E.R.E., 8-9 Bedford Sq., London W.C.1)

BRIGHTON

Mar. 2–6 Exhibition Halls Engineering Design Show (Business Conferences & Exhibitions, Mercury House, Waterloo Rd., London S.E.1)

CAMBRIDGE

Mar. 19–22 Churchill College Television Tomorrow (Royal Television Society, 166 Shaftesbury Ave., London W.C.2)

CRANFIELD

Mar. 23–26 College of Aeronautics Aerospace Instrumentation Symposium (N. O. Matthews, Dept. of Flight, College of Aeronautics, Cranfield, Beds.)

EDINBURGH

Mar. 17-20 Management and Economics in the Electronics Industry (D. J. T. Williams, Ferranti Ltd., Ferry Rd., Edinburgh 5)

HARWELL

Apr. 2-3 **High Voltage Electron Microscopy** (1.P.P.S., 47 Belgrave Sq., London S.W.1)

MANCHESTER

- Jan. 6-8 The University Solid State Physics (1.P.P.S., 47 Belgrave Sq., London S.W.1) Feb. 23-27 Belle Vue Labex Northern
- (U.T.P. Exhibitions, 36-37 Furnival St., London E.C.4)
- May 19–22 Belle Vue Industrial Training Exhibition & Symposium (John Clarke (P.R.) Ltd., St. James House,
 - 44 Brazennose St., Manchester 2)

OXFORD

Apr. 6-11 The University Biological Engineering Conference (J. Gasking, Dept. of Pharmacology, St. Bartholomews Hospital Medical School, Charterhouse Sq., London E.C.1)

Wireless World, January 1970

Sept. 14–16 The University Photo-electron Spectroscopy (I.P.P.S., 47 Belgrave Sq., London S.W.1)

READING

- Apr. 6–8 The University Thin Films Conference
- (I.P.P.S., 47 Belgrave Sq., London S.W.1) Apr. 15–17 The University Defects in Semiconductors
 - (I.P.P.S., 47 Belgrave Sq., London S.W.1)

TEDDINGTON

Feb. 25-26 N.P.L. Trends in Diffusion Conference (I.P.P.S., 47 Belgrave Sq., London S.W.1)

UXBRIDGE

Apr. 14–16 Brunel University Computer Graphics International Symposium (R. Elliot Green, Brunel University, Uxbridge, Middx.)

Overseas

JANUARY-APRIL lan. 14-16 Honolulu System Sciences Conference (Dr. R. H. Jones, 2565 The Mall, University of Hawaii, Honolulu, Hawaii 96822) Ian. 20 & 21 Chicago Soldering Technology Seminar (W. R. Dunbar, Grover M. Hermann Hall, Illinois Inst. of Technology, 3241 S. Federal St., Chicago, Illinois 60616) Paris Feb. 6-11 Audiovisual Techniques, Electroacoustics & Electronics Show (Fed. Nat. des Ind. Electroniques, 16 rue de Presles, Paris 15) Feb 16-19 Tampa Fla. Computer-Aided Circuit Optimization Dr. G. W. Zobrist, Dept. of Elect. Eng., University of South Florida, Tampa, Florida 33620) Feb. 18-20 Philadelphia Solid-State Circuits Conference (I.E.E.E., 345 E. 47th St., New York, N.Y. 10017 Feb. 24-Mar. 5 Tampa, Fla. **Applied Communication Systems Analysis** (Dr. G. W. Zobrist, Dept. of Elect. Eng. University of S. Florida, Tampa, Fla. 33620) Mar. 5-10 Paris **Audio Festival** (Fed. Nat. des Ind. Electroniques, 16 rue de Presles, Paris 15) Mar. 11-13 Washington Scintillation and Semiconductor Counter Symposium (Louis Costrell, Radiation Physics Inst. Section, N.B.S., Washington, D.C. 20234) Mar. 18-21 Nairohi Electro 70 Show Electronics Institution of East Africa, P.O. Box 9690, Nairobi, Kenya) 3-8 Paris Apr. Electronic Components Show (Fed. Nat. des Ind. Electroniques, 16 rue de Presles, Paris 15) Apr. 6-10 Paris Advanced Microelectronics Conference (Fed. Nat. des Ind. Electroniques, 16 rue de Presles, Paris 15) 14-17 Washington Apr. Geoscience Electronics Symposium (I.E.E.E., 345 East 47th St., New York, N.Y. 10017) Apr. 21-24 Budapest Microwave Communication Colloquium (Microcoll—Technica Háza Budapest, V. Szabadsag tér 17, Hungary) 27-29 Atlantic City Frequency Control Symposium (Electronic Components Lab., Electronics Command, Fort U.S. Army Monmouth.

New Jersey 07703)

Wireless World, January 1970



General Purpose Audio Amplifier

The Elcom GPA general purpose amplifier module-available in two different supply voltage versions - is a low-power highquality amplifier intended for rack mounting and capable of driving a speaker to 4W mean power. Output is unbalanced, but an external transformer is available for balanced loads. Power supplies required are balanced loads, rower suppres required are 50V for the GPA50 and 24V for the GPA24. The a.c. coupled output has an impedance of 7.5 % (GPA24) and 15 %(GPA50). The transformer coupled output has an impedance of 3 9, 7.5 9 and 15 9 for both voltage versions. At 4W output, sensitivity is quoted as 20dBm maximum, noise level 100dB, and distortion 0.3% (GPA24) and 0.1% (GPA50). Frequency response from 30Hz to 20kHz is within ± 1 dB, and the amplifier is provided with remote or local gain control facilities. Elcom (Northampton) Ltd, Weedon Road Industrial Estate, Northampton. WW 310 for further details.

Auto-range Digital Voltmeter

The SM 212/C digital voltmeter from Laboratories is designed specifically S.E. for data logging and the automatic testing of equipment systems. Selecting "auto" on



the front panel gives complete automatic continuous monitoring over the full range of the instrument (from 10 N to 1kV). The instrument up-ranges at a reading of 9,000 and down-ranges at a reading of 800. All control switches are front-mounted push-buttons. As well as having automatic selection, the instrument's ranges can be selected manually by five push-buttons, and remote ranging for external programming via the B.C.D. output socket. Maximum resolution is 10 µV, input impedance > 1000M Ω on direct ranges, and accuracy

±0.01%. Reading rate is 25 per sec. synchronized to mains frequency. Seriesmode rejection is >60dB without filter, common-mode rejection > 140dB. and S.E. Laboratories (Engineering) Ltd, North Feltham Trading Estate, Feltham, Middx. WW 315 for further details.

Turns-counting Dial

R. C. Knight Ltd introduce a range of turnscounting dials for use with multi-turn potentiometers. The model 33-30 illustrated can be used on 10-, 20-, or 30-turn potentiometers. Mating with the potentiometer shaft is accom-



plished using a tapered collet arrangement. Small quantities of up to 100 off are available from stock and prices range from 41s 9d to 47s 3d each, depending on the quantity ordered. R. C. Knight Ltd, 20 Solent Avenue, Lymington, Hants.

WW 336 for further details

Constant-current Power Sources

Direct currents as small as 1 nA and as large as 500mA are supplied with extreme accuracy by two constant-current power sources-models 6177B and 6181Bfrom Hewlett-Packard. Current regulation is such that the output current changes less than 25 p.p.m. (±5 p.p.m. of range setting) with a load change that swings the output voltage from zero to maximum. Current-setting and voltage-limiting controls are independent and can be preset before the load is connected. Maximum output for Model 6177B is 500mA, with voltage limiting continuously adjustable between 0 and 50V. Model 6181B has a maximum output of 250mA with 0-100V limiting. Current output is selected with high resolution (0.2% of range) by a 10-turn control and 3-position ×10 range switch. Either of the floating output terminals may

www.americanradiohistory.com



be grounded to provide current of either polarity. For systems use, these instruments can be programmed by either external voltage or resistance changes. Extremely high output impedance is maintained without use of reactive elements, resulting in fast programming speed: 500 s from 0 to 99% of programmed output. Hewlett-Packard Ltd, 224 Bath Road, Slough, Bucks.

WW 316 for further details

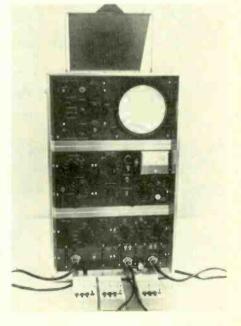
Light-emitting Diode

The MV50 from Monsanto, is a diffused planar gallium arsenide phosphide lightemitting diode which peaks at 6,500Å. It can be used in place of incandescent lamps as small as the T3/4 size. The life-time of the unit is said to approach 100 years. Light output is 750ft lamberts with a forward current of only 20mA. Turn-on time is 1ns. The MV50 is available in the U.K. from Semiconductor Specialists Inc, Airpark House, 127 Station Road, West Drayton, Middx.

WW 330 for further details

Electromyograph

Isleworth Electronics have developed an electromyograph suitable for use in clinical medicine practice. The type 7 Electromyograph is a solid-state design, with plug-in units. Four plug-in positions are provided: two amplifiers, a timebase and a stimulator. Two types of amplifier are available-a single channel and double channel. With various combinations, between two and four signals may be displayed simultaneously on the cathode-ray tube. Each channel carries its own built-in calibration signal. Comprehensive camera recording facilities are

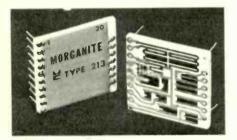


provided by a second cathode-ray tube at the rear of the instrument. The remote cable-operated camera shutter is synchronized electrically with the machine. Five modes of operation are possible: single sweep, continuous, superimpose, scan, and autograph. All modes are electrically interlocked to prevent overlapping exposure. Film wind-on after each exposure is automatic. Outputs are provided for data recorder, external loudspeaker (to supplement the unit's own internal sound channel), and trigger pulses to synchronize peripheral equipment. Isleworth Electronics, Frederick Street, Waddesdon, Bucks.

WW 320 for further details.

Binary Ladder Networks

Morganite Resistors have produced two binary ladder networks. Model 215 is an eight-bit high-speed ladder with a standard resistance of $10k \Omega$ in the binary R-2R configuration. Settling time is less than 100ns with a maximum output voltage ratio error of ± 0.25 bit over the temperature range -55 to $\pm 125^{\circ}$ C. Three application resistors, ratioed to the ladder, are also included for bipolar operation and ampli-



fier summation. Model 213 is a 10-bit binary resistor array designed to be compatible with the Fairchild μA 722 current source for D/A and A/D applications. Resistance values correspond directly with those specified on the A 722 data sheet. The resistance temperature coefficient is 0 to -200 p.p.m./°C with resistance tolerance of $\pm 2\%$ over the temperature range -20 to ± 85 °C. Morganite Resistors Ltd, Bede Industrial Estate, Jarrow, Co.

Durham. WW 334 for further details.

Analogue Computer Teaching System

Analogikit is a system which combines digital and analogue techniques and is intended for the designer, technician and university graduate. The equipment includes operational amplifiers, feedback, summing, scaling and initial condition elements. An introductory handbook explains the arithmetical aspects of summing, scaling and integration with simple illustrative experiments leading to the construction of differential equations with practical examples of simulated systems. An advanced book deals with the mathematical aspects for degree course students. A complete teaching kit for analogue work, including a mounting deck with power supplies, costs £165 in the U.K. Feedback Ltd, Park Road, Crowborough, Sussex. **WW 308 for further details.**

W W 508 for further details.

Automatic Counter

All models in the Dana series 8100 automatic counters measure frequencies from 0.05Hz to 50 MHz (d.c. coupled) and 5Hz to 50MHz (a.c. coupled), but models 8120, 8130, 8124 and 8134 have an additional a.c. coupled range from 10 MHz to 500MHz. Models 8110, 8130, 8114, 8134 have a facility for time interval measurement in the range 0.1 s to 10s (up to 100s as an option). When using one of these automatic counters, all that is necessary is to connect the input signal, and set an input voltage range switch to PRESET. The remainder of the measurement process is then controlled by computer logic. Automatic resolution during the reading time is effected, and decimal points and units are automatically indicated. In one second, reading accuracy approaches $\pm 2 \times 10^{-7}$, It is claimed the accuracy of the 8100 is better without operator adjustment than that of a manually controlled counter which requires function selection and periodto-frequency calculation. Dana Electronics Ltd, Bilton Way, Dallow Road, Luton, Bedfordshire.

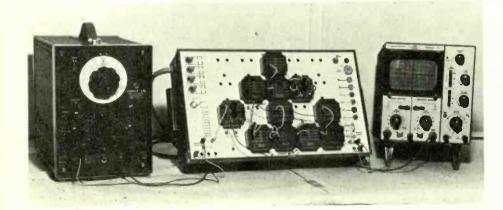
WW 305 for further details.

32-MHz Counter

The GR 1192 counter measures time intervals, frequency and period from d.c. to 32MHz to a resolution of 0.1 μ s, and frequency ratio. If the 1157-B scaler is used, the frequency range can be extended to



500MHz. Models are available with 5, 6 or 7 digit presentation. When measuring frequency from d.c. to 32MHz, the counting gate times are $100\mu s$ to 10s, and Hz, kHz and MHz can be displayed with positioned



decimal point. Accuracy is ±1 count \pm time base accuracy. The stability of the 10MHz time base is less than ± 2 parts in 10⁻⁶ per month. Measurement of period is limited by the digit presentation and is up to 100s, 10s and 1s in the 7-, 6- and 5-digit models, respectively. Single and multiple periods up to 10⁵ are covered, and time periods can be displayed in ms, µs and ns with positioned decimal point. The ratio of two frequencies, A and B, can be measured from 1 to 105. Frequency A from d.c. to 32MHz is measured over 1 to 10⁵ periods of frequency B, 50Hz to 10MHz. Trigger error in time measurements is defined at $\pm 0.3\%$ of one period divided by the number of periods averaged for a 40dB input signal-to-noise ratio, and assumes no noise internal to the counter. For input signals of extremely high signal-to-noise ratio, the trigger error in us is less than 0.0003 divided by the signal slope in $V/\mu s$. Price of the three models is £326 (5 digit); £382 (6 digit) and £437 (7 digit). General Radio Company (U.K.) Ltd, Bourne End, Buckinghamshire.

WW 303 for further details.

3mm Jack Receptacle

Sealectro have developed a 3mm s.r.m. series jack receptacle which prevents r.f. radiation. Designated part number



50-645-4520-31, the receptacle is constructed of gold plated stainless steel, Teflon, and gold-plated beryllium copper. It meets all requirements of MIL-C-39012 regarding contact and dielectric torque and captivation. R. F. Components Division, Sealectro Ltd, Walton Road, Farlington, Portsmouth, PO6 1TB.

WW 317 for further details.

DC/DC Converters

A range of d.c./d.c. converters for changing an available low d.c. voltage (between six and sixty volts in multiples of six volts) to a much higher d.c. voltage is available from Plessey. General use of the component is in transistor instruments incorporating a cathode-ray tube for display purposes. Normal voltages up to 8kV with power up to 0.5W are available in the unit size of 12.7 \times 5.1 \times 3.8cm, and ranging up to 10kV and 2W in a unit measuring 13.4 \times 5.9 \times 4.5cm. Plessey Wound Components Division, Titchfield, Hants. WW 321 for further details.

Portable Instrumentation Recorder

A portable instrumentation magnetic tape recorder, especially designed for use by nonskilled personnel, has been introduced by Bell & Howell. Suitable for a wide range of industrial and research applications, the VR-3200 is available in both 4-track and 6-track versions with speed ranges of $1\frac{2}{8}$ to 15 i.p.s. and $3\frac{1}{4}$ to 30 i.p.s. respec-



tively. Both versions use $\frac{1}{4}$ in tape on a $10\frac{1}{2}$ in diameter standard N.A.B. spool. F.M. circuitry gives a frequency response of d.c. – 10kHz at 30 i.p.s. with a signalnoise ratio of 44dB. Recording and playback are possible at any speed, the correct centre frequency and filter being automatically selected from the tape transport speed selection switch. Peak detection monitor meters are fitted which enable the correct modulation to be adapted for any input signal level in the range 100mV to 30V. A test facility is incorporated to allow setting up without running tape. Operation is from 230V a.c. (50Hz). A remote control unit is available as an optional extra. Price from \pounds 1,456. Bell & Howell Ltd, Consolidated Electrodynamics Division, Lennox Road, Basingstoke, Hants.

WW 325 for further details.

Marine Communications Receiver

Marine general-purpose/s.s.b. communications receiver type R551, from Redifon, is designed to meet the British Post Office specification and international requirements for a ship's main and s.s.b. receiver. All-solidstate it provides unbroken frequency coverage from 60kHz to 30MHz. It incorporates a frequency synthesizer as well as continuously variable of "free" tuning.



Frequency of the basic receiver is set on three in-line direct-reading decade dials which select the digits for MHz tens, MHz units and kHz hundreds (e.g. 27.2MHz). The tuning process is then completed by a continuously variable control which sweeps a range 100kHz wide. This control is geared to a counter in the same line as the decade dials to give a direct read-out of frequency down to 100Hz (e.g. 27.2796MHz, ог 27,279.6kHz). Complete frequency synthesis, using the ARU11, provides the additional decades to enable operators to select precisely the digits for kHz tens, kHz units, and Hz hundreds. This permits immediate direct setting of known frequencies down to increments of 0.1kHz, with instant switch selection of free tuning or of complete synthesis. Another feature is a dynamic range of over 120dB. This accommodates a correspondingly wide variation in levels of input signals. The use of two inter-coupled a.g.c. systems enables the R551 to receive wanted signals at sensibly constant level despite adjacent unwanted signals. A high degree of front-end protection, independent of whether the receiver is switched on or off, permits the R551 to be installed and operated in close proximity to high-power transmitters. Redifon Ltd, Broomhill Road, London S.W.18. WW 312 for further details.

S.S.B. Manpack

The TRA.6929 Minical s.s.b. manpack from Racal-BCC has a power output of 1W p.e.p. and six operating channels covering fre-quencies from 2 to 7MHz or 2.6 to 9MHz. Complete with batteries, handset, aerial and haversack, the manpack measures 190 \times 76 \times 210 mm and weighs 3.6 kg.



Intended for military use, Minical has been specifically designed for simplicity of operation with a minimum number of controls. Changing channel takes only a few seconds, and it is claimed that unskilled operators are able to use it after a few minutes instruction. Tuning is effected with the aid of an internal noise generator. By this means, the possibility of radiation during tuning is avoided. The manpack operates from either U2/D type dry cells or rechargeable NiCad cells. If required, a vehicle 12V supply can be used to power the manpack or for recharging the NiCad cells. Racal-BCC Ltd, Western Road, Bracknell, Berkshire.

WW301 for further details

R.F./U.H.F. Millivoltmeters

Millivac r.f. millivoltmeters types MV-828A and MV-928A are solid-state instruments with full-scale ranges from 1mV to 3V, extended to 300V by a capacitative divider. Frequency range is 10kHz to 1,200MHz. Both instruments operate on 115/230V, 50-450Hz supplies, whilst the MV-928A also operates on internal nickel-cadmium batteries and has a built-in battery charger. Features of the instruments include temperature-compensated probe with replacement diode cartridge and recorder output. Avail-



through Millivac's U.K. agents, able Lyons Instruments Ltd, Hoddesdon, Herts, the MV-828A is priced at £337 10s and the MV-928A at £387 10s. (Duty free). WW 326 for further details.

Integral-cycle Zero-voltage Switch

The CA3059 integral-cycle zero-voltage switch is contained within a 14-lead dual in-line plastic package and operates direct from the a.c. line. This RCA device is capable of driving triac gates directly, and by providing a triac gating signal at zerovoltage crossings minimizes r.f. inter-ference. A fail-safe circuit is incorporated to guard against an accidentally opened or shorted sensor, and an optional output control is available. Electrical characteristics include; d.c. gate-trigger current of 40mA for V_{GT} of 3V and R_{GT} of 70 Ω , gate-trigger pulse width of 80 us before and after 'O' for $C_{\chi} = 0$; a gate-trigger pulse width of 20us before 'O' and 170µs after 'O' for C x of 0.015µF; an on-off accuracy of 1% and 3% for sensors of $5k\Omega$ and $100k\Omega$, respectively. RCA Ltd, Sunbury-on-Thames, Middlesex. WW 309 for further details.

Dual-trace 10-MHz Oscilloscope

The D54 is a solid-state-circuit oscilloscope. Intended for general purpose laboratory and production line testing applications the Telequipment D54 solid-state oscilloscope has a vertical amplifier bandwidth of d.c. to 10 MHz within -3dB when d.c. coupled,



and 2Hz to 10MHz within -3dB when a.c. coupled. A 12-position frequency compensated input attenuator calibrated direct in V/cm can be set for sensitivities of 10mV/cm to 50V/cm in a 1-2-5 sequence. The range of timebase sweep speeds is 200ns/cm to 2s/cm covered in 22 calibrated steps. A variable uncalibrated control provides continuous overlap between steps and reduces slowest sweep speed to approximately 5s/cm. The D54 can be operated in the following four modes: channel 1 only; channel 2 only; alternate during which the input to the vertical output amplifier is synchronously switched between channel 1 and channel 2 during flyback; and chopped during which the input to the vertical output amplifier is continuously switched between channel 1 and channel 2 at approximately 100kHz. Telequipment Ltd, 313 Chase Road, Southgate, London N.14.

WW 302 for further details.

Over-voltage Protection Unit

New from ITT is an over-voltage protection unit designed to give semiconductor devices protection against voltage surges of 1 µs or greater duration. The unit employs a reference amplifier and variable potential divider to sense applied voltage. The trip point is continuously variable between 4.5 and 60V, with resolution better than 0.1V. An excess voltage triggers a crowbar

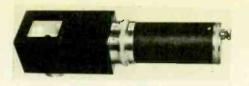


s.c.r. across the supply. In the event of a fault the unit will handle 500A peak-250A mean half cycle. Provision is made for limiting the surge current to lower values if desired. Connected across the two supply terminals, the unit takes less than 10mA drain at all voltages. The unit is compactapproximately 65 × 40 × 50mm. Access to the voltage adjustment potentiometer is through a hole adjacent to the terminals. The unit will operate in an ambient temperature range from -40° to $+65^{\circ}$ C. ITT Components Group Europe, Rectifier Product Division, Edinburgh Way, Harlow, Essex.

WW 319 for further details.

Colour Monitor Calibrator

The Grafikon calibrator is a hand-held optical instrument that enables the white point of a colour monitor or receiver to be visually set. The instrument is offered up to the tube face and the monitor controls are adjusted to make the colour picture match the instrument's reference colour. The comparison between the monitor and the reference is seen in a Lummer Brodhun photometer cube. The reference colour is obtained from a tungsten halogen lamp and glass filter and is diffused to form a very even reference field. The lamp current is electronically stabilized to ensure that its col-



our is the same each time it is switched on while the tungsten halogen cycle in the lamp maintains its long-term colour stability. A mechanical iris is incorporated to adjust the brightness of the reference field to any grey scale step. This ensures that the reference colour remains the same for all values of brightness. Grafikon Engineers Ltd, 75 South Western Road, Twickenham, Middx. WW 331 for further details.

Transportable Insulation Tester

Miles Hivolt offer a transportable insulation test set for measuring leakage currents down to 0.01 A at up to 30kV d.c. It can be driven from the mains or from rechargeable 24V batteries which give four hours' use at full load, the equivalent of many days' normal use. Mains input is 100-125V or 200-250V at 45-66Hz. The output voltage is available in two ranges 0.5-5kV and 3-30kV. The output voltage is measured at two ranges 5kV and 30kV with full-scale deflections. Maximum output current is 200 A at full voltage with higher currents at lower voltages. The equipment weighs 11.5kg with either battery or mains power unit fitted. Miles Hivolt Ltd, Riverbank Works, Old Shoreham Road, Shoreham-by-Sea. Sussex.

WW 313 for further details.

40-MHz Counter/Timer

A solid-state 40-MHz counter/timer, the TF2414A, with 10mV sensitivity is avail-able from Marconi Instruments. Advantages include time interval measurement down to 1 µs, period and multi-period measurement, 1M / input impedance and display memory. Direct frequency measurement is provided up to 40MHz. A special version TF2414A/2M is designed for use with the M.I. frequency converters (TF2400 series) which extend the frequency range up to 500MHz. Another version, TF2414A/1 provides a printer output facility supplying a 1-2-4-8 b.c.d. output code for each digit displayed. Stability and accuracy are determined by an oven-controlled crystal oscillator. Circuits incorporate discrete and integrated silicon semiconductors on plug-in printed boards. The display memory maintains the readout while the count is in progress, thus giving a continuous coherent readout. The crystal oscillator has stability of typically 1 \times 10⁻⁶ over three months and temperature co-efficient of $\pm 5 \times 10^{-7}$ per °C. A standard frequency output is available from the internal reference



built-in monitor, has a Y bandwidth of d.c.

to 100kHz. The instrument costs £24. 10s.

(discount for schools) and features a Y sensitivity of approximately 100mV/cm at maximum gain with full Y shift. The timebase range covers approximately 100ms/cm to 10µs/cm and is automatically synchronized. The X input required is 1V/cm with full X shift when the timebase is switched off. Flyback suppression and access to Y plates-

oscillator through a front panel socket over a range from 0.1Hz to 1MHz (selected by the range switch). Price £298 f.o.b. U.K. Marconi Instruments Ltd, Longacres,

WW 311 for further details.

St Albans, Herts

Add-on Transmitter Amplifier

A wideband untuned solid-state linear transmitting amplifier covering the frequency range 2-30MHz, intended as an add-on unit to a low-power transmitter /receiver such as a packset, has been developed by The M.E.L. Equipment Company Ltd. It has an output of 100W p.e.p. and is designed for a transmitter, the

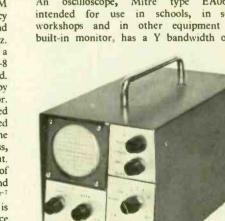


output of which is not less than 5W p.e.p. It operates from d.c. supplies of 10-30V without need for voltage adjustment. The power supply is self-contained, and provides a supplementary output of 20W at 24V d.c. for the associated packset. It is sealed, operable at ambient temperatures from -15° C to $+55^{\circ}$ C and meets the durability requirements of DEF 133 (L3). Designated type BA.1013/01 the amplifier measures 310mm wide by 300mm deep by 116mm high. The M.E.L. Equipment Company Ltd, Manor Royal, Crawley, Sussex

Educational oscilloscope

WW 335 for further details.

An oscilloscope, Mitre type EA0699-1, intended for use in schools, in service workshops and in other equipment as a



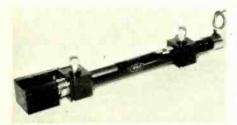
Wireless World, January 1970

are also featured. Power supply required is 25W at 200 to 250V, 50 to 60Hz. Mitre Electronic Products, 22 Powis Terrace, London W.11.

WW 304 for further details.

Broad-band Travelling-wave Amplifier

The TWS23 travelling-wave amplifier from the M-O Valve Company has an output of greater than 10W flat to better than 3dB and at a gain of at least 26dB in the frequency range 2.0 to 4.0GHz, whilst, by adjusting helix voltages, it is possible to



obtain power outputs of over 20 watts at spot frequencies. The tube is packed in a permanent magnet focusing mount, and r.f. coupling is by means of a type N 50 connector. Cooling may be by either conduction or convection, according to specification. The M-O Valve Co. Ltd, Brook Green Works, London W.6. WW 322 for further details.

F.E.T—input Op. Amp.

Advance Industrial Electronics have available a low-cost general purpose Zeltex operational amplifier, Model 134D which can be used in differential, inverting and non-inverting circuits. Typical voltage drift is 50 µV/°C. Typical input bias current is 50pA. Initial offset voltage can be adjusted to zero with an external potentiometer. The unit is short-circuit proof to ground The amplifier housed in a



plastic /epoxy case can be mounted directly on a p.c. board or plugged into a mating connector. Specification includes an output of $\pm V$ at 4mA, a d.c. gain of 50,000, and a gain/ bandwidth product of 1.3 MHz. Frequency at full output is 100kHz. The slew rate is 6V/s and operating temperature -25 to +85°C. Price is £8 5s. Advance Institute Electronics, Raynham Road, Bishops Stortford, Herts

WW 333 for further details

V.L.F. Third-octave Analyser

AIM Electronics have announced a thirdoctave frequency analyser (TOF 260A) with a frequency range extending from 0.5Hz up to 100kHz and covering any eight octaves in this frequency range. The octaves

covered are pre-set to customer requirements. The unit consists of twenty-four filters, each covering one-third of an octave, designed in accordance with BS2475:1964 (which recommends centre frequencies and equivalent bandwidth of the filter elements). Each filter may be attenuated by 0-100% by adjustment of a ten-turn calibrated potentiometer. The outputs from all the filters are combined at the output socket. Thus any combination of filters may be selected by adjustment of the attenuators. Typical applications include extraction of third-octave information from unknown waveforms and simulating the characteristic noise of any low-frequency excitation (e.g. vibrations) by selective filtering of white noise. Price £600. AIM Electronics Ltd, Bar Hill, Cambridge, CB3 8EZ. WW 323 for further details.

Transistor Amplifier for 1-2 GHz

Electro/Data Inc. have developed a broadband transistor amplifier for the range 1 to 2 GHz. The new amplifier, designated Model A-12, has a 15dB gain response from 1 to 2 GHz with greater than 10dB of gain from 700 MHz to 2.2 GHz. The



amplifier's noise figure is 6dB typical with a maximum value of 8dB. It has miniature 50 ? input and output connectors and a shielded d.c. bias input. A single, negative 12V, 14mA source is required for biasing. Two or more units can be cascaded to provide increased gain, with minor changes in passband ripple and bandwidth. The amplifier has linear gain for output signals up to -10dBm. Electro/Data Inc., 1621 Jupiter Road, Garland, Texas 75040, U.S.A.

WW 314 for further details.

Modulation Meter

Type 785 modulation meter from Dymar is a solid-state instrument for the measurement of the depth of modulation in a.m. transmitters or the frequency deviation in the case of f.m. transmitters. It is specifically designed for narrow deviation transmitters in mobile and portable v.h.f. radiotelephones, the most sensitive deviation range being 3kHz f.s.d. The frequency range covered is 30-480 MHz and the sensitivity over the whole of this range is better than 2.5mV in 50 (-40dBm) which permits loose coupling to the transmitter under test. The residual f.m. noise of the local oscillator is typically -44dB below 3kHz deviation with



the a.f. "voice" filter switched in. Auxiliary outputs are provided at the i.f. (500kHz) and the demodulation audio frequency. This permits viewing of the modulation waveforms on an oscilloscope or applying it to distortion analyzers. Price £240. Dymar Electronics Ltd, Colonial Way, Radlett Road, Watford, Herts

WW 327 for further details.

Analogue Switches

A range of m.o.s.a.i.c. analogue switches, the ML150 series, has been introduced by Plessey. The switches, with full gate-control isolation and gate-oxide protection, are available in 6-way multiplexer, dual sample/hold and 3-bit digital-to-analogue configurations. The MP130 series provide matching drive circuits for ML150. The large negative output voltage swings of these circuits (30V) are particularly suited to driving m.o.s. analogue switches. Plessey Microelectronics, Cheney Manor, Swindon, Wilts.

WW 324 for further details.

Continuous Tape-**Transport System**

A continuous magnetic tape-transport system-MTD 10500-is announced by Recording Designs Ltd. The system comprises three basic models, write only, read only and write /read. Each has variants to suit a range of requirements. The same tape-transport technique is used for each with modular electronics to give particular system characteristics. Seven- and nine-track versions are available each with bi-directional transport speeds from 4 to 37.5 i.p.s. as standard, and an optional speed-range of between 1 and 75 i.p.s., if required. Slew-mode speed (for highspeed inter-block gap detection) is 120 i.p.s. Start and stop speed times are less than 20ms in the standard speed range. Recording densities of 200, 556 and 800 b.p.i. are available. Recording Designs Ltd, Blackwater Station Estate, Camberley, Surrey.

WW 332 for futher details.

Low-pass Active Filters

Lionmount are manufacturing low-pass active filters which can be varied continuously throughout the passband. Two types are available; one of which covers the range 1 to 10kHz in one band; the other covering the frequency range 1Hz-10kHz in four switched bands. The designs are based on



9th order Butterworth or Chebychev configuration and can realize 80dB/decade attenuation at cut off. The filters will accept an input voltage of ±10V peak and may be loaded with a minimum of 2,000 (2 Lionmount & Co. Ltd, Bellevue Road, New Southgate, London N.11. WW 318 for further details.

January Meetings

Tickets are required for some meetings: readers are advised, therefore, to communicate with the society concerned

LONDON

6th. I.E.E. "Marketing and the component engineer" by R. H. W. Burkett at 17.30 at Savoy Pl., W.C.2.

6th. I.E.R.E .- Discussion on "The Haslegrave report on technician courses and examinations" at 18.00 at the London School of Hygiene and Tropical Medicine, Keppel St., W.C.1.

7th. R. Soc.—Juvenile lecture "Television at school" by Dr. R. C. G. Williams at 14.30 at John Adam St., W.C.2. 7th. L.E.R.E.—"Positional transducers and

precision electronic measurements" by P, C. F.

Wolfendale at 18.00 at 9 Bedford Sq., W.C.1. 9th. 1.E.E. "Tellegen's Theorem: an unusual theorem of wide circuit application" by Dr. R. Spence at 17.30 at Savoy Pl., W.C.2. 12th. 1.E.E.—"Phasor diagrams". Discussion led

by M. G. Scroggie at 17.30 at Savoy Pl., W.C.2. 13th. I.E.E.—Discussion on "Prospects for ultra-high-frequency f.e.ts" at 17.30 at Savoy Pl., W.C.2.

13th. 1.E.R.E. /I.E.E. "Physiology for engineers -control of circulation" by Dr. I. Gabe at 18.00 at St. Bartholomew's Hospital Medical Coll., E.C.1.

14th. I.E.E.—"Changing relations between science and technology and their effect on international co-operation" by Dr. A. P. Speiser at 17.30 at Savoy Pl., W.C.2. 14th. I.E.R.E.—"U.H.F. television transposer

equipment" by W. L. Gregory at 18.00 at 9 Bedford Sq., W.C.1. 20th. I.E.E.—"The British Calibration Service"

20th. 1.E.E.—"The Brush Calibration Service" by H. E. Barnett at 17.30 at Savoy Pl., W.C.2. 20th. 1. Electronics.—"Power semiconductor electronics" by R. G. Dancy at 18.30 at the London School of Hygiene, Keppel St., W.C.1. 21st. S.E.R.T.—"The new I.V.C. colour video tape recorder" by R. A. Calaz at 19.00 at London

School of Hygiene, Keppel St., W.C.1. 22nd. J.E.E.—"Devices using tunnelling super-currents" by Dr. B. D. Josephson at 17.30 at

Savoy Pl., W.C.2. 23rd. Brit. Acous. Soc.—Symposium on "Electroacoustics in air and water" at 10.00 at 1

Birdcage Walk, S.W.1. 26th. I.E.E.—"Satellite television distribution" by A. K. Jeffris, D. G. Pope and P. C. Gilbert at 17.30 at Savoy Pl., W.C.2. 28th. I.E.R.E.—Colloquium on "Systems

engineering and its educational impact" at 18.00 at 9 Bedford Sq., W.C.1. 30th. I.E.E.—"Radar ecnoes from clear air in

relation to refracting-index variations in the troposphere" by J. A. Lane at 17.30 at Savoy Pl., W.C.2.

BELFAST

21st. I.E.R.E.—"Air traffic control" by David Evans at 18.30 at Ashby Institute, Queens University, Stranmillis Road.

BIRMINGHAM

26th. I.E.E./I.P.O.E.—"Operational experience with p.c.m. systems" by D. Cleobury at 18.00 at M.E.B., Summer Lane.

BOLTON

12th. I.E.E.T.E .- "The origins of electrical communications" by J. Dalton at 19.30 at Institute of Technology, Deane Rd.

BRISTOL

15th. I.E.R.E. /R.Ac.S. /I.E.E .- "B.A.C. satellites" by G. Crowder at 19.00 at Filton House Conference Room, Filton.

CAMBRIDGE

29th. I.E.R.E./I.E.E.-"Tuning of gunn effect oscillators" by P. W. Crane at 18.30 at University Engineering Laboratories, Trumpington Street.

CARDIFF

14th. I.E.R.E.—"Electronics for process control instrumentation" by J. Seers at 18.30 at University of Wales Institute of Science and Technology. 16th. S.E.R.T.—"Educational use of C.C.TV" by T. Evans at 19.30 at College of Further

Education, Cyncoed.

22nd. R.T.S.—"Television transmission equip-ment in education" by W. D. Kemp at 19.00 at B.B.C., Llandaff.

СНАТНАМ

15th. I.E.R.E.—"The engineer in management" by F. Oakes at 19.00 at Medway College of Technology.

CHELMSFORD

19th. I.E.R.E./I.E.E .- "Radar ornithology" by Dr. E. Eastwood at 18.30 at the Civic Centre, Duke Street.

EDINBURGH

7th, I.E.R.E.—"Pulse code modulation for point-to-point music transmission" by E. Rout at 19.00 at Napier College of Science and Technology, Colinton Road.

20th. I.E.E.—"The electronics industry in Scotland—past, present & future" by I. MacDonald at 18.00 at the Carlton Hotel.

FARNBOROUGH

22nd. I.E.E. /I.E.R.E .- "Speech and vocoders" by L. C. Kelly at 19.00 at the Technical College.

GLASGOW

8th. I.E.R.E.—"Pulse code modulation for point-to-point music transmission" by E. Rout at 19.00 at the Institution of Engineers and Shipbuilders, 183 Bath St., C.2.

LEEDS

6th. I.E.E.-""The automatic landing of aircraft" by S. A. W. Jolliffe at 18.30 at the University.

LEICESTER

20th. I.E.R.E.—"Static inverters and their applications" by E. W. Porter and R. J. Green at 18.30 at the University.

29th. I.E.E.—"The latest techniques in mputer-aided electronic design" by E. computer-aided Wolfendale at 18.30 at the City Polytechnic.

LIVERPOOL

5th. 1.E.E.-"Communications for people at ork and at play" by D. G. Holloway at 18.30 at the University.

7th. I.E.E. (Grads.)—"Low and medium frequency noise in transistors" by Dr. K. F. Knott at 18.30 at M.A.N.W.E.B. Elec. Indus. Development Centre.

19th. 1.E.E.-"Electronics in automobiles" by W. G. Hill at 18.30 at the University.
 22nd. 1.E.E. (Grads.)—"Developments in radio

control" by J. R. Francis and R. T. King at 18.30 at M.A.N.W.E.B. Elec. Indus. Development Centre.

29th. 1.E.E.—Faraday Lecture "People, communications and engineering" by J. H. H. Merriman at 10.15 and 14.30 (students) and 18.45 (public) at the Philharmonic Hall.

MANCHESTER

7th. I.E.E./I.E.R.E.-"On the future of world communications" by Prof. C. Cherry at 18.15 at U.M.I.S.T

28th. I.E.E. (Grads.)-"Radio interference from high-voltage transmission line conductors" by M. G. Faulkner at 18.45 at U.M.I.S.T.

NEWCASTLE-UPON-TYNE

7th. S.E.R.T.—"Tandberg audio" by A. W. Dakin at 19.30 at Charles Trevelyan Technical College, Maple Terrace.

14th, I.E.R.E.—"Electronic telephone exchanges" by V. E. Mann at 18.00 at Dept. of Physics and Physical Electronics, Rutherford Coll., Ellison P1.

14th, I.E.E.T.E.-"Decca navigational system" by A. Brooker-Carey at 19.30 at Rutherford College

of Technology, Ellison Place. 26th. I.E.E.—"The application of electronic engineering to road safety" by D. G. W. Mace and S. Penoyre at 18.30 at the Polytechnic.

NEWPORT

21st. I.E.E.T.E .- "Change to metric" by G. Esplin at 19.30 at College of Technology, Allt-Yr-Yn Avenue.

NOTTINGHAM

13th. I.E.E.-Faraday Lecture "People, communciations and engineering" by J. H. H. Merriman at 14.30 (students) and 19.15 (public) at Albert Hall.

OXFORD

14th. 1.E.E.—"Tomorrow's world—use of satellites for communication" by W. J. Bray at 19.00 at College of Technology, Headington.

PRESTON

14th, I.E.E.-"'Metrication" by T. C. Campbell at 19.30 at Yorella Restaurant.

READING

22nd. I.E.R.E.-"M.O.S. devices in I.s.i." by G. E. Stevenson at 19.30 at J. J. Thomson Physical Laboratory, the University, Whiteknights Park.

20th. I.E.E. (Grads.)—"Brain cell to microcircuit (pattern recognition)" by Dr. I. Aleksander at 18.15 at the College of Engineering Technology.

SOUTHAMPTON

27th. S.E.R.T .- "Field effect transistors" by G. A. allcock at 19.30 at the College of Technology, East Park Terrace.

28th. 1.E.R.E.-"Electronic character recognition" by R. H. Britt at 18.30 at the Lanchester Theatre, University.

STEVENAGE

12th. I.E.E.—"Current electronic developments in the deep sea fishing industry" by P. J. Hearn at 19.30 at the College of Further Education.

STOKE-ON-TRENT

15th. 1.E.E.—Faraday Lecture "People, communications and engineering" by J. H. H. Merriman at 14.30 (students) and 19.30 (public) at Victoria Hall, Hanley.

SUNDERLAND

22nd. I.E.E. (Grads.)-"Pulse code modulation" by J. Hutton at 18.30 at the Polytechnic.

WEYMOUTH

29th. I.E.E.—"Applications of integrated circuits" at 18.30 at South dorset Technical College.

WOLVERHAMPTON

7th. I.Prod.E. /I.Mech.E.—"C.E.I. as a professional union" by K. M. Platt at 19.15 at Stafford College of Technology, Beacon Side.

Test Your Knowledge

Series devised by L. Ibbotson*, B.Sc., A.Inst.P., M.I.E.E.,

M.I.E.R.E.

20. Colour

In all the questions it is assumed that the viewer has normal colour vision.

1. Select from the colours quoted below the one which does not appear in the spectrum of white light:

- (a) orange
- (b) yellow
- (c) purple
- (d) violet.
- (-)

2. From the spectral colours below select the one which is associated with the highest frequency of radiation:

- (a) red
- (b) blue
- (c) green
- (d) blue-green.

3. Three light sources of the same area give monochromatic radiation of colours red, green and blue respectively, and have equal luminosity (appear equally bright). The intensity of radiation:

- (a) is the same for all three
- (b) is least for the red
- (c) is least for the green
- (d) is least for the blue.

4. Evidence suggests that the human brain distinguishes between different colours by the relative stimulation of optical receptors having different frequency responses, in the eye. The theory is that:

- (a) each "cone" in the retina has a frequency response curve which is slightly different from all the others
- (b) a separate type of receptor responds to each spectral colour
- (c) only three distinct frequencyresponse characteristics are involved
- (d) only two distinct frequencyresponse characteristics are involved

5. Monochromatic light of wavelength $580m\mu$ is seen as yellow. It therefore follows that any light entering the eye which appears to have the same hue:

- (a) must consist of monochromatic light of wavelength 580mµ
- (b) may contain many frequencies, but must have maximum energy flux at 580mu

*West Ham College of Technology, London E.15.

- (c) must contain some energy at wavelength 580mµ, but not necessarily have maximum energy flux at this wavelength
- (d) need not contain any energy at 580mu wavelength

6. True white light is:

- (a) light with equal energy at all frequencies in the visible range
- (b) light with a spectral distribution the same as that emitted by the sun
- (c) the light emitted by a "black body" at a temperature of 5200°K
- (d) an inexact concept which is defined differently in different circumstances.

7. A single monochromatic light can be rendered colourless (giving the sensation of white) by the addition of a suitable quantity of another monochromatic light:

- (a) whatever the colour of the original light
- (b) unless the original light is in the red region of the spectrum
- (c) unless the original light is in the green region of the spectrum
- (d) unless the original light is in the blue region of the spectrum.

8. White light falls on an object which absorbs in the blue, but reflects other frequencies. The colour of the object will be seen to be:

- (a) yellow
- (b) green
- (c) red
- (d) purple.

9. Monochromatic yellow light from a sodium lamp falls on an orange (fruit). The colour of the orange when viewed in this light will be:

- (a) very pale orange
- (b) low intensity orange
- (c) yellow
- (d) black.

10. If white light is added to light of any given colour the result is:

- (a) a change in hue, but no change in saturation of the colour
- (b) a change in saturation, but no change in hue
- (c) a change in both hue and saturation

(d) if the original light was monochromatic a change of saturation only, otherwise a change of both hue and saturation.

11. A particular green light has a radiant flux density of 1 watt per square metre. To this light is now added 1 watt per square meter of pure violet light. The effect will be:

- (a) a considerable change in colour, but little change in luminance
- (b) a large increase in luminance, but little change in colour
- (c) little change in either colour or luminance
- (d) a large change in both colour and luminance.

12. Discounting luminance information, the colour of a light can be specified entirely using:

- (a) one variable
- (b) two variables
- (c) three variables
- (d) seven variables.

13. If three colours are located on the chromaticity diagram, then mixtures of varying (positive) quantities of light of these three colours will produce only:

- (a) all colours within the spectral locus (all realisable colours)
- (b) all colours inside the triangle having the given three colours at the corners
- (c) all colours outside the triangle having the given three colours at the corners
- (d) all colours on straight lines joining the three given colours.

14. By mixing, in appropriate quantities, fully saturated red, green and blue light it is possible to produce light:

- (a) of all colours (all hues and saturations)
- (b) of every hue, but not all saturations
- (c) over a restricted range of hues, but with all saturations in that range
- (d) over a restricted range of both hues and saturations.

15. If ideal phosphors could be developed which produced monochromatic red (700mu), green ($520m\mu$) and blue ($450m\mu$) light, these could be used, with advantage, at the output of a colour television system. The camera filters at the input of the system would require:

- (a) to pass bands of frequencies, as narrow as possible, at the quoted wavelengths
- (b) to have broad overlapping frequency transmission characteristics with maximum transmissions at the quoted wavelength values
- (c) to have pass-bands which met but did not overlap, so as to divide the visible spectrum into three bands centred on the three quoted wavelengths
- (d) to have pass-bands between the quoted wavelength values.

Answers and comments, page 47

Literature Received

ACTIVE DEVICES

"The use of Coaxial-Package Transistors in Microstripline Circuits" is the title of Application Note AN-4025 which has been published by RCA Electronic Components, Harrison, New Jersey 07029, U.S.A.**WW401**

Dickson Electronic Corp's field-effect and bipolar transistors in l.i.d. chip assemblies are described in a 19-page brochure which may be obtained from Dage (Great Britain) Ltd, 1 Penn Place, Rickmansworth, Herts. WW403

Ferranti Ltd., Gem Mill, Chadderton, Oldham, Lancs, have produced some additions for their Microspot c.r.t. manual. This includes a contents sheet and provisional data on the types IB /97, 12H /40, 14 /08, 16A /19, 16A /40, 21B /10 cathode-ray tubes, the DY605 electronic display equipment and the PD5002 solid-state light source.WW405

The semiconductor products of SGS (UK) Ltd, Planar House, Walton Street, Aylesbury, Bucks, are listed in two catalogues which are available price 21s each.

Consumer devices. Professional discrete devices

PASSIVE COMPONENTS

Airpax Electronics, of Cambridge, Maryland 21613, U.S.A., have produced the following two leaflets.

Programming systems produced by Oxley Developments Co., Priory Park, Ulverston, Lancs, are the subject of a new catalogue WW414

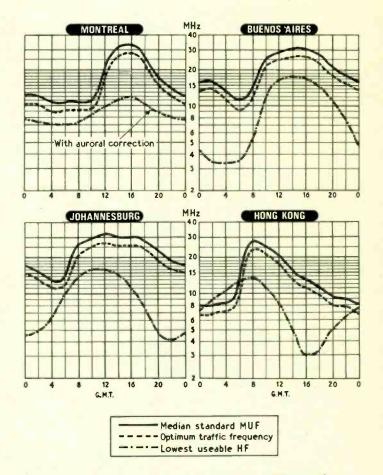
"Professional Communications Antenna Systems" is the title of a catalogue available from Antenna Specialists UK Ltd, 1 Euston Road, London N.W.1

An article entitled "Understanding Thermocouples" that originally appeared in our sister journal *Instrument and Control Engineering* has now been reprinted and is available from IPC Business Press (Sundry Sales Department), 161–166 Fleet Street, London E.C.4. Price 6s 9d, including postage.

EQUIPMENT

An effects generator which can be used to produce sound effects for radio and TV programme inserts, and public address announcements is described in a leaflet from the manufacturers Mellotronics Ltd, 28–30 Market Place, London W.1 We have received the following publications from Marconi Instruments Ltd, Longacres, St. Albans, Herts.

H.F. Predictions—January



Winter season conditions will continue with a large differential between day and night frequencies except on some routes which show a secondary peak a few hours before dawn. At sunrise and sunset, therefore, the rate of change of MUF is at its greatest and it becomes difficult to maintain satisfactory communication over these periods. On shorter routes, generally less than 2000 km, the daytime MUFs in winter may be lower than in summer when propagation is via the E layer.

The LUFs shown were calculated by Cable and Wireless Ltd for reception in the United Kingdom of point-to-point telegraph services. For other services the curves would be displaced vertically, the exact amount depending on service and equipment parameters.

Answers to **''Test Your** Knowledge''

Questions on page 45

THE CHOICE

1. (c) No single monochromatic radiation produces the sensation purple; it requires a mixture from the two ends of the spectrum. If monochromatic red light of wavelength 700 mµ and monochromatic violet of wavelength 400 mµ (the normally accepted ends of the spectrum) are mixed in various proportions, then the range of "pure" purples is produced.

2.(b)

3. (c) Since the eye is most sensitive to light in the green part of the spectrum, far less energy-flux density is required to produce a given luminance of green light than is required for the same luminance of red or blue. Note that the term "monochromatic" is used to describe radiation of one single wavelength (or, in practice, since this is impossible, over a very narrow band) even though in the present context it may seem inappropriate.

4. (c) The details of the mechanism are still not known.

5. (d) Suitable quantities of light of other wavelengths can produce a similar stimulus in the colour receptors. If spectral green and red are used the result will have the same saturation as well as the same hue.

6. (d) The lights described in (a), (b) and (c) are all forms of white light, although their spectral energy distributions are somewhat different. The standard of white used in television is the colour of a light produced by a particular combination of a tungsten lamp and a filter, known as "standard illuminant C'

7. (c) The complementary of green is purple, which is non-spectral.

8. (a) Yellow is the complementary of blue, so that removing blue from white leaves yellow.

9. (c) The orange, like most natural coloured objects, reflects light over a range of wavelengths, so that its colour, when illuminated by white light, is determined by the total effect of these on the eye. Since the orange is here illuminated with pure yellow it can only reflect yellow.

10. (b) This, and most other properties of colour, are well illustrated by the chromaticity diagram.

11. (a) Although wavelengths at the blue end of the spectrum contribute very little to the brightness of a light they have a very significant effect on its colour.

12. (b) These can be dominant wavelength and purity, or chromaticity co-ordinates (as in the chromaticity diagram). This is why colour information in colour television can be carried by two signals.

13. (b)

14. (b) Reference to the chromaticity diagram shows that no triangle with its corners on realisable colours, even on the spectral locus, can include the whole diagram.

15. (b) The total transmission characteristics for each of the three colour channels in the camera would require to be such that any incident radiation would produce relative responses in the three channels equal to the relative outputs from the three phosphors at the receiver required to produce its colour as nearly as possible.

BULGIN PREGISION COMPONENTS PIONEERS FOR NEARLY 50 YEARS AND TODAY LEADING THE WORLD OF CRITICS IN NEW ESSENTIAL COMPONENTS The Bulgin policy of continued research and development has resulted in the introduction of many unique new Electronic Components during the past year. A few are illustrated here. B, 16, 17, 18 three-panel mounting Battery Holders accepting 1, 2, or 3 U2-sized cells respectively. SM. 257/2 + K. 515. Semi-rotary shaft operated D.P.C.O. moulded body switch rated 2A. at 250V A.C. D/S. 941/1 and /2. Illuminated Switch with a normally biased push action which can be locked in the depressed position. L.E.S. lamps, single (/1) or twin (/2) S.P.C.O. switch unit. SM. 301/2/PD moulded body, D.P.C.O. 8-contact switch for double-pole alternative circuit switching. Rated 3A. at 250V A.C. P. 537/Chrome or /Gold. Three pole side entry, BS. 666, jack plugs. K. 556/Legend. Collet fixing knob, dial and escutcheon unit which can have dial P. 537/Chrome legending to customers' requirements. or/Gold SM. 301/2/PD D/S. 941/1 & /2 SM. 257/2 + K. 515 Knob D/S. 890/SA. 2419 F. 316/S D. 965, 966 K. 556/Legend Example of Edge Legendin F. 296/S F. 317, 318 D/S. 890/SA. 2419 Switched. P. 550 Legended Indicator. L.E.S. Lamps, with a choice of five lens colours which can be legended. Switching is D.P.C.O. push-push successional action rated 2A. 250V A.C. Other models have different switching arrangements. D. 965, 966 New L.E.S. Signal Lamps for direct connection to printed circuit boards with choice of SM. 324/2 five lens colours, transparent or translucent, F. 316/S Panel mounting fuseholder for 1" x ‡" ø fuses 15A, rating, F. 296/S Miniature panel mounting fuseholder for 5 x 20 mm, fuses, 5A. rating. SM. 324/2 Key operated D.P.C.O. moulded switch 2A. 250V A.C. rating. P. 550 unique 7 pole + earth inlet/outlet connector rated 6A. 250V A.C. F. 317, 318 Flush-fitting panel mounting fuseholder 5A. rating, F.317 1" x $\frac{1}{4}$ ", F. 318. 1 $\frac{1}{4}$ " x $\frac{1}{4}$ " fuses. Edge Legending. All transparent knobs. K. 436-7 and 472-4 can have legending around the edges as well as on top surface. FOR DETAILS OF THE COMPLETE RANGE SEND FOR BROCHURES REF. W.W./1. AANUFACTURERS AND SUPPLIERS OF RADIO A. F. BULGIN & CO. LTD., AND ELECTRONIC COMPONENTS TO IRALTY MINISTRY OF WORKS Bye Pass Rd., Barking, Essex. MINISTRY OF AVIATION MINISTRY OF TECHNOLOGY RESEARCH ESTABLISHMENTS U.K.A.E.A. WAR OFFICE AIR MINISTRY Tel: 01-594 5588 (12 lines) E OF

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WN AGENTS

Real & Imaginary

by Vector

"Yellow, and black, and pale and hectic red,"

As I write, the persuaders have just begun their honeyed blandishments in the Press, on sound radio and on television.

Like those purposeful citizens who make a crust by robbing strongrooms, the colour vendors use an oblique approach. Just as cracksmen traditionally begin operations in the cellars of the house next door to the bank, so do our persuaders tunnel into your private strongroom at your weakest point, namely the Little Woman and to some extent the kids, because these, as a generalization, look at the box a lot more than you do.

At present several of the channels are hard at it, backed by powerful newspaper and magazine campaigns—'it' in this case being the task of making you feel a second-class citizen if you are still viewing in unnatural monochrome.

Every day now, and far into the night, the B.B.C. and I.T.A. are firing continual salvoes extolling the merits of colour. What does astonish me is that, at the time of writing, I haven't seen any advertisment emanating from a radio manufacturer on any of the channels.

Should any reader be reaching for his pen to remind me that the I.T.A. is the only organization permitted to carry advertising, stay your hand. While it is true that the B.B.C. does not lend itself to the sordid business of raking in money in return for advertising time, there are other, and more gentlemanly, ways of going about it, as any press relations officer worth his salt could tell you.

One such is for the would-be advertiser to latch on to some national sport or cult. One of your first acts is to present a handsome trophy which has the name of your product indissolubly attached to it; all you have to do then is to sit back and wait for the event to be televised.

Another method is to plaster the railings of the more dynamic association football clubs with advertisements of your product. Try as he may, the cameraman will have to have the railings in the picture for a good deal of the time and so, given a little luck, you have a free plug both on B.B.C. and I.T.A. for about nine months in any given year.

I see the B.B.C. is making a platform of 'natural television'. They did something similar some years ago in a drive to popularize the v.h.f. sound service. 'High quality' was the torch carried then, but this was soon extinguished by the radio manufacturers, who shoved cheese-pared circuits into a small box, together with a tinpot loudspeaker, and tried to sell it as hi-fi.

And what is natural television, pray? If the term means anything it signifies that, colour-wise, the picture on the home screen is identical with the scene in the studio. That being so, I must say that I'm surprised that the B.B.C's technical boys have allowed their advertising colleagues to get away with it. For, given an additive system with all its registration problems, the inclusion of band-saving techniques, and colour filters with transmission characteristics which only approximate to those of the home receiver's phosphors, then even the best colour monitors will not stand comparison with the actual scene.

It's a pity to have selected such a sales story because it looks as if the experience of v.h.f. sound is going to be repeated with colour television, if the criminally maladjusted receivers to be seen all too often in dealers' showrooms are anything to go by.

The public will swallow it of course for the same reason that, in 1922, it subscribed to the belief that an unbiased three-valve receiver feeding a 'sugar-loaf' horn loudspeaker was giving perfect quality. They believed it because the only standard of comparison was the acoustic gramophone and the quality of the wireless' was, in its day, marginally better than that. By the same token, today's standard of comparison with a colour set is the monochrome receiver and therefore any colour, however unreal, is better than no colour at all. Provided that the sky is some shade of blue and the grass approximates to green, who cares about fidelity? Electronics engineers, certainly, and artists, perhaps, but precious few else. So there was really no need for the B.B.C. to oversell on fidelity.

Sticking my neck into the prophet's noose, my guess is that colour television will take several years to become the norm in the average home and that not a few manufacturers will catch colds in the process.

What, I wonder, would happen if

someone came up with a colour system that was miles ahead of PAL? If there is any such lone inventor reading this, I would advise him that he is most unlikely to see his brain-child come into general use. For, with about $\pounds 150M$ already invested in the present system, nobody is going to look kindly upon an invention that sets everyone back to square one.

What are the prospects of such a happening? Who knows? What would it be like, this super system? This also is anybody's guess. Almost certainly, I would think, it would embody a subtractive colour system. It would also employ a translating interface which is much more in accord with the human eye-brain complex than is today's television camera.

Our present system is a hangover from Clerk Maxwell, who was the first to show that three black-and-white transparencies can, under certain conditions, provide a picture in full colour. This is an application of the Young-Helmholtz trichromatic theory which is generally believed to form the basis of human colour vision in spite of some anomalies which cannot easily be explained away. No one, for instance, has positively identified three types of cone structure in the eye, one red-detecting, one green and one blue; all the cones seem pretty much the same. Then, a few years ago Dr. Edwin ("Polaroid") Land demonstrated that two colours, or even one red light and one white, can interact to provide a gamut of colour. Even two monochromatic light sources will produce a wide variety of diluted colours. (This in fact was no new discovery; colour film processes, using two colours only, have been patented since the turn of the century.)

There is a growing awareness of the extreme complexity of the human eye and it is possible that further discoveries in this area may provide the electronics industry with important new thinking about television. We are moving away from comfortable concepts where, for instance, 500 mµ always equates with green light to quicksands where a body radiating at $500m\mu$ appears to the eye as brilliant red. (Yes, I know it sounds daft but it can be done by interfering with the signals which trigger the brain into registering colour.) These coded signals are the core of the matter; if only the code could be broken, all sorts of possibilities exist. It might even be feasible to dispense with conventional displays and, instead, feed signals to the area behind the retina.

This sounds crackpot until we come to terms with the thought that colour sensations needn't derive from incident light-frequency radiation. The coded signals to the brain can be affected in various ways; by mechanical vibration; by the application of external voltages or currents or by hallucinatory drugs. In the last-mentioned case manifestations occur which have every semblance of three-dimensional reality. Given an exact control of the input signals, what might not be possible? Even a degree of sight to the blind seems to be feasible. why the It's still an Avometer yet fits in the pocket/held easily in one hand Has a d.c. sensitivity of 10,000Ω/V Measures up to 25kV and 25A with optional accessories Accuracy conforms to B.S.S. 89/54. ni_ ter me

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WW-006 FOR FURTHER DETAILS

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* High-stability VFO of 2 FET's and 2 transistors and easily handles QSO's for hours. * Precision double gear dial—a TRIO innovation —with linear frequency variable capacitor. Possible to get finer reading 1KHz. One dial rotation covers 25KHz, makes SSB demodulation easier. * Frequency range covers entire amateur band from 3.5MHz to 29.7MHz. One-touch selection system switches bands. WWV reception of 15MHz possible. * MHz band circuit structure patterned on Collins type double conversion system so first oscillation is by crystal control, second local oscillation by VFO.

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- * FREQUENCY RANGE: 3.5-29.7 MHz (7 Bands)
- * SENSITIVITY: 1µV (at 10 dB S/N)
- * IMAGE RATIO: More than 50dB
- * FREQUENCY STABILITY: ±2KHz in 1–60min. after
- switching on, subsequently within 100 Hz per 30 min.
- * Dimensions: 13"(W), 7-3/32"(H), 12-3/16"(D).



Model SP-5D COMMUNICATION SPEAKER

- * Communications Speaker which has been designed for use with the 9R-59DE.
- Dimensions: 3-9/16"(W), 7-1/8"(H), 5-3/16"(D).



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4CX1000A 4CX1000K	_	1.0	3.2	3.0	110	6.0	9.0
4CX1500B		1.5	2.7	3.0	30	6.0	9.0
4CX5000A	CV8295	5.0	16.0	7.5	30/110	7.5	75
4CX10,000D	CV6184	10.0	16.0	7.5	30/110	7.5	75
4CX35,000C	_	35.0	82.0	20.0	30	10	300
CR192A (6166A)	CV8244	10.0	9.0	6.9	60/220	5.0	175
Vapour Cooled	Anode dissipation	Output	Anode voltage	Frequency	Filament ra	tings	Boiler
Туре	max. (kW)	(kW)	max. (kv)	(MHz)	(∨)	(A)	unit
CY1170J	60	82	15	30	10	300	Integral
CY1172 (RS 2002V)	150	220	15	30	21	350	CY4120



4CX1000K

For audio or linear single sideband amplifiers. 4CX1000K has a solid disc screen contact to permit use up to 400MHz.

Send for full details of EEV tetrodes



4CX10,000D

or screen

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For audio, linear,

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4CX35,000C

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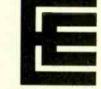




CY1172

English Electric Valve Co Ltd Chelmsford Essex England Telephone: 61777 Telex: 99103 Grams: Enelectico Chelmsford

For audio amplifiers, r.f. linear amplifiers or Class C amplifiers or oscillators. Both types have a coaxial metal-ceramic envelope. A range of glass envelope types is also available.



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Please send me full data on your range of forced-air cooled and vapour cooled tetrodes. I am also looking for a power tetrode with the following parameters. Output Anode voltage Frequency

power (kW)	max (kV)	(MHz)	
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COMPANY			
ADDRESS			
TELEPHONE NUMBER			
TEEET HONE NOWBER		EXTENSION	10/10/20

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AP 358

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kW

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COMPANY			
ADDRESS			
TELEPHONE NUMBER	EXTENS	SION	WW4U
WW-010 F	OR FURTHER DETAILS		AP 354

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	D	9	3	6.5		46/6	45/6	44/6	4/6	
	E	9	7	6.5		49/6	48/6	47/6	4/6	
	F	9	10	6.5		58/6	57/6	56/6	4/6	
	G	13	3	6.5		49/6	48/6	47/6	4/6	
	H	13	7	6.5		58/6	57/6	56/6	4/6	
	1	13	10	6.5	E	69/6	68/6	67/6	6/-	
	J	18	3	6.5		58/6	57/6	56/6	4/6	
	К	18	7	6.5		79/6	77/6	76/6	6/-	
	L	18	10	6.5	- T	106/-	104/-	103/-	6/	
	M	4.5	3	13		36/6	35/6	34/6	4/6	
	N	4.5	7	13		49/6	48/6	47/6	4/6	
	0	4.5	10	13		69/6	68/6	67/6	6/-	
	P	9	3	13		49/6	48/6	47/6	4/6	
	Q	9	17	13		69/6	68/6	67/6	6/-	
	R	9	10	13		79/6	77/6	76/6	6/	
	S	13	3	13		58/6	57/6	56/6	6/-	
	Т	13	7	13		79/6	77/6	76/6	6/	
	Ų	13	10	13		99/6	98/-	97/-	7/6	
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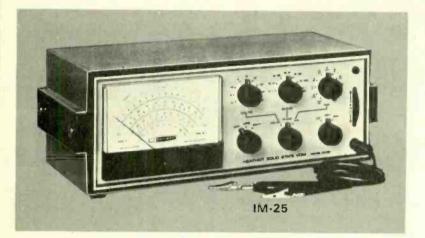
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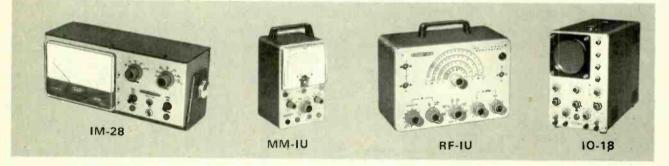
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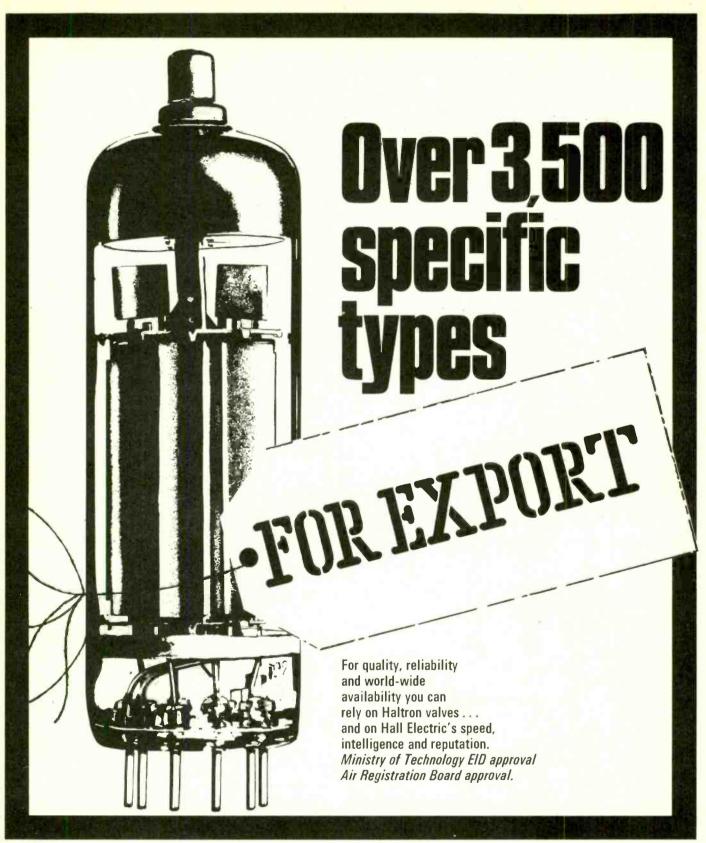
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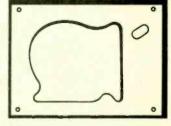


MODEL 2000 PLINTH SYSTEM

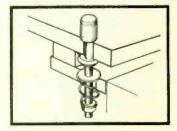
The SME model 2000 plinth system is more than a handsome and convenient housing for your turntable and SME precision pick-up arm. It meets the mechanical requirements under which the best performance will be obtained. High-quality workmanship is combined with ease of assembly. The basic unit is finished in selected veneers of teak, straight-grained walnut, or rosewood. A one-piece hinged lid in heavy acrylic is reinforced with a polished stainless-steel trim.



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New Buyers' Guide

There's a new wallchart on Mullard special quality receiving valves. It gives comprehensive equivalents information, and it's free from any Mullard Industrial Distributor, or just use the reader enquiry service.



Tektronix Type 556 DC-to-50 MHz, dual-beam, sweep-delay oscilloscope

The Type 556 and rack-mount Type R556 use any combination of Tektronix letter or 1-series plug-ins

The UPPER BEAM can display a signal from either *left* or *right* plug-in; with either Time Base A, Time Base B, or external signals; triggered from a composite vertical signal, plug-in single channel signal (with 1A1 or 1A2), external, or line.

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Independent Sweep Systems provide 24 calibrated steps from $0.1 \,\mu$ s/cm to 5s/cm; the X10 Magnifier extends the fastest sweep rates to 10ns/cm.

Calibrated Sweep Delay extends continuously from 0.1 microsecond to 50 seconds.

Signal-Sweep Operation enables oneshot displays of normal or delayed sweeps.

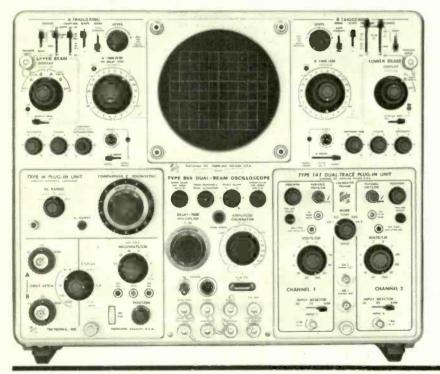
Independent Triggering Systems provide

The LOWER BEAM can display a signal from the *right* plug-in; with either Time Base B or external signals; triggered from a composite vertical signal, plug-in single channel signal (with 1A1 or 1A2), external, or line.

stable displays to beyond 50MHz. Either input signal can be used to trigger either or both time-bases.

New Dual-Beam CRT (with illuminated internal graticule) provides "zero-parallax" viewing of small spot size and uniform focus over the 8cm by 10cm display area. Each beam has 6cm vertical scan, with overlap scan of 4cm by 10cm.

EMI (RFI) Suppression — meets interference specifications of MLL-I-6181D over these frequency ranges : 150kHz to1GHz — Radiated (with CRT mesh filter installed), and 150kHz to 25MHz — Conducted (power line).

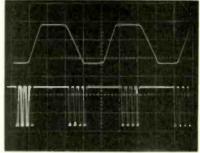


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Simultaneous Single-shot Displays. Current versus voltage display of a 0.75 ampere, fast-blow fuse during destructive overload. Both beams are driven by B Time-Base at 50 µs/cm which is delayed by pre-triggered A Time-Base to provide base reference lines before and after the event. The upper beam shows the current waveform at 30A/cm while the lower beam shows the corresponding voltage across the fuse at 100V/cm.



Single-input Dual-beam Displays. Upper beam shows bursts of 2.5MHz pulses on Time Base A with time variation between bursts. This shows up as increasing time-jitter between the first and successive bursts. The lower beam shows B Sweep (0.1 μ s/cm) delayed by A Sweep and triggered on the second pulse of the last burst to provide a jitter-free expanded display of the A Sweep intensified zone. The use of only one probe and one plug-in input simplifies signal connection and provides minimum loading on the source.

Plug-in units shown

Type 1A1 Dual-trace Unit (Dual-Trace—50mV/cm at DC-to-50 MHz, 5mV/cm at DC-to-28 MHz. Single-Trace—500 μ V/cm at 2Hz-to-15 MHz. 5 Display Modes—Channel 1, Channel 2, Alternate, Chopped, Added Algebraically, Front-panel signal out put.) Type W Differential

Comparator Unit

(Conventional Preamplifier—50mV/cm at DC-to-23MHz to 1mV/cm at DC-to-8MHz. Decade Input Attenuator to X1000. Differential Input Preamplifier— CMRR of 20,000 to 1, DC-to-20kHz. Max Peak Input of \pm 15V, XI Attenuation. Calibrated Differential Comparator —Vc Supply of 0 to \pm 11V. Accuracy of \pm 0.15% of output \pm 0.05% FS.)

Price without plug-in units-£1,560

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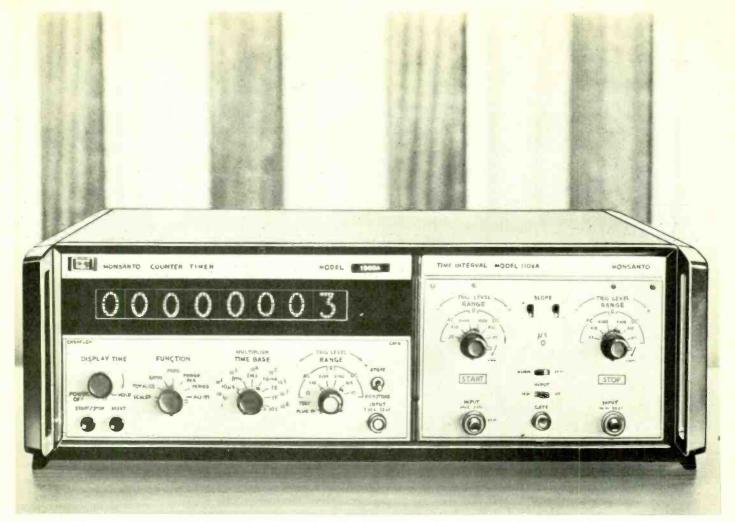


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Average Accuracy 0.1%. Residual resistance: 4 Decade - 12 milliohms, 5 Decade - 15 milliohms.

Decade	X 100,000 ohms	X 10.000 ohms	× 1.000 ohms	× 100 ohms	× 10 ohms	X1 ohm	<0.1 ohm
Max. Cont. Current	3 mA	7 mA	20 mA	70 mA	200 mA	700 mA	1A
Short time current rating	3.2 mA	10 mA	32 mA	100 mA	320 mA	1 A	1 A

Precision Metal film decade resistance box suitable for use both on d.c. and at high frequencies. Accuracy actually improves with use and age.

5 Decade Cat. Ref. R7 1,111,100 Ω by 10 Ω steps Cat. Ref. R9 111,110 (by 1 Ω teps Cat. Ref. R10 11,111 Ω by 0.1 Ω steps 4 Decade Cat. Ref. R5 111,100 by 10 Ω steps Cat. Ref. R4 11,110 by 1 Ω steps Cat. Ref. R3 1,111 by 0.1 Ω steps

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Air Cored Decade Inductor Accuracy ±5%

- 3	decade	model	Cat.	Ref.	L1	1	mΗ	to	1	Henry	
2	decade	model	Cat.	Ref.	12	1	mΗ	to	10	00 mH	
2	decade	model	Cat.	Ref.	L3	10	mΗ	to	1	Henry	
AL.	o available					-					

Also available precision Air Spaced Capacitors. Conductance Boxes. Loading Resistors, standard Resistors and Capacitors. Potentiometers, Resistance Bridges and Solid State d.c. Null Detectors.

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SPECIFICATIONS

SFECIFICATIONS		-				
	TYPE TA401	TYPE TA601	ТҮРЕ ТА605			
GAIN	40dB ±0.1dB	60dB ±0.1dB	20, 30, 40, 50 and 60dB ±0.2dB.			
BANDWIDTH ±3dB	1 Hz-3MHz	3Hz-1.2MHz	20-40dB, 1Hz-3MHz; 50dB, 2Hz-2MHz; 60dB 4Hz-1.5MHz.			
BANDWIDTH ±0.3dB	4Hz-1MHz	10Hz-300kHz	20-40dB, 4Hz-1MHz; 60dB, 10Hz-300kHz.			
INPUT IMPEDANCE	>5M Q, <40pF from 100Hz to 1MHz	>IM Ω , <50pF from 100Hz to 300kHz	>5M Ω . <40pF from 100Hz to 300kHz.			
INPUT NOISE		<15 μ V, zero source; <40 μ V, 100k Ω source	As TA401 and TA601 at 40dB and 60dB.			
POWER SUPPLY	PP3 battery,	life 100 hours	PP9 battery, life 1,000 hours, or A.C. Power Unit			
AVAILABLE OUTPUT	IV up to IMHz. 300mV 100kΩand 50pF	at 3MHz. into load of	1.5V up to 2MHz. IV at 3MHz. into 100k Ω and 50pF			
OUTPUT IMPEDANCE		100 <i>Ω</i> in se	eries with 6.4 μ F			
SIZE AND WEIGHT	3″ x 1≩″ x	1 1 ″ 7 oz.	2½" x 4" x 5½" 2½ lb.			
PRICE with Battery and input lead	£17.0.0 £17.0.0		£27.0.0 (Optional A.C. Power Unit £7.10.0 extra).			

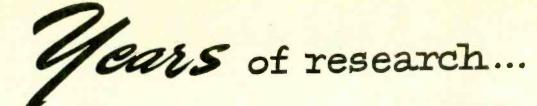


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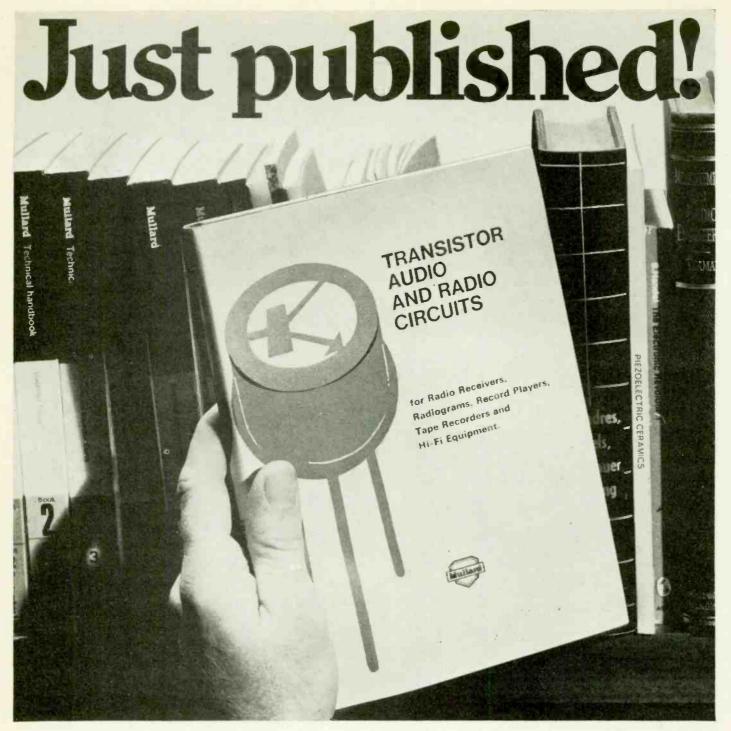
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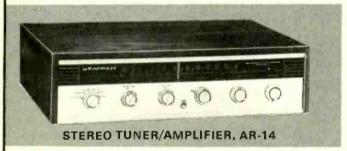


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STEREO 'COMPACTS'

The newest additions to the Heathkit range are two "stereo compacts". The AD-27, pictured left includes a turntable unit with a Shure magnetic cartridge, an FM stereo tuner and a 30 watt stereo amplifier. The whole is built into an attractive compact teak or walnut veneered cabinet - all for a kit price of only £82! The AD-17 compact is similar but does not have the FM radio facility and uses a simpler but still attractive cabinet. This kit only costs £54.

STEREO TUNER AMPLIFIERS

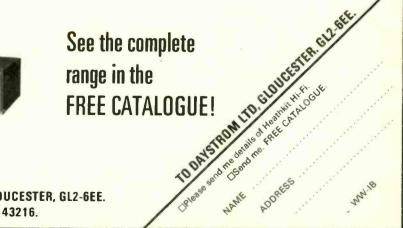
If you need a Tuner-Amplifier, we can offer models to suit any pocket. Pictured on the left is the very popular Heathkit AR-14. This is a solid-state stereo Tuner-Amplifier with a sensitive FM tuner, a built-in stereo decoder and a 30 watt stereo amplifier (15 watts I.H.F.M. per channel). It is wonderful value at a kit price of £54.

STEREO 'SEPARATES'

If your preference is "separates", or perhaps you want just a stereo amplifier without a tuner, again Heathkit offers a selection. Typical is the TSA-12 stereo amplifier, illustrated. This is a solid state stereo amplifier (15 watts I.H.F.M. per channel) at a kit price of only £32 16 0. We have radio tuners to match either for FM reception only, or for FM and Long and Medium wave. The Stereo Tuner, model TFM-15 costs only £28 14 0 in kit form.

LOUDSPEAKER SYSTEMS

All the units described above can be used with any good hi-fi loudspeakers. To cover this need, the Heathkit range includes several hi-fi loudspeaker kits. The Berkeley kit features a 12 in. bass loudspeaker and a 4 in. highfrequency unit, a ready finished teak or walnut veneered cabinet, and the kit price is only £21 4 0. The 'Avon' mini kit is only £13 8 0.



WW-024 FOR FURTHER DETAILS

audix portable D.J. Console

Specification

Power Output: 25-30 watts at 100V line. Inputs: Two microphones 30-50 g balanced. Auxiliary switched gram, tape, or medium impedance microphone.

Mister Gain Control, Tape Record Socket, Bass Tone Control: + 10 db - 12 db @ 100 cps. Treble Tone Control: + 8.5 db - 10 db @ 10 Kc/s. Frequency Response: ± .5 db 25 c/s - 12 Kc/s. Dimensions: Length: 3' 4''. Depth: 1' 6''. Height: 2' 6''. Height less legs: 1' 0''. Meets the ever growing demand for well presented music and speech at private and public functions—dances, receptions, parties etc. where there is no permanent sound installation. The Audio Portable D.J. Console incorporates the A.25 silicon transistorised power amplifier with three input channels each with separate volume control. The two Garrard SP.25 single play turntable units each have continuously variable fader controls. Monitor loudspeaker, socket for headset and dynamic cardioid microphone completes the equipment which is attractively housed in solid afromosia cabinet with detachable legs.

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SOUND

SYSTEMS

Goldring 800 magnetic cartridges track unerringly. Because that's the way we make them. They're designed to translate even the most delicate information stored in the groove back into an identical electrical signal. We call it the sound of true transduction.

Hear it for yourself. You'll know you're on the right track.

Goldring 800/H... the 800/H is designed for inexpensive changers to track between $2\frac{1}{2}-3\frac{1}{2}$ grams and has a high output of at least 8mV. £10.13.6 tax paid. Goldring 800

the 800 is designed for standard arms and changers where the requirements of high fidelity and robustness usually conflict. £13.0.0 tax paid.



Goldring 800 E... is designed for transcription arms and a micro-elliptical diamond is fitted to a finer cantilever, end damped against natural tube resonance £18.17.1 tax paid. Goldring 800 Super E... the 800 Super E is for those to whom perfection is barely good enough. Extraordinarily low mechanical impedance for superb tracking capabilities. Each cartridge is supplied with its individual curve and calibration certificate. £26.0.0 tax paid.

Send for details and complete range of Goldring Hi-Fi equipment Goldring Manufacturing Co. (Great Britain) Ltd. 486-488 High Road, Leytonstone, London, E.11. Tel: C1-539 8343. www-028 FOR FURTHER DETAILS Wireless World, January 1970





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GHS, GHD SERIES – DUAL PURPOSE For laboratory or industrial use. Capable of achieving ultimate pressures down to 5 x 10-4 And giving a 50% increase in pumping speed over previous models.

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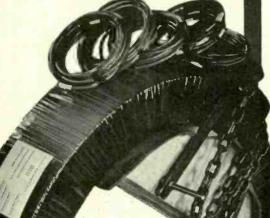
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Telcon Mumetal used in a large bushing current transformer, manufactured by Smith Hobson Ltd., which was one of several for the Niagara Falls Generating Station.



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Pye Telecommunications is the world's largest exporter of radiotelephone equipment. Pye Radiotelephones are used all over the world to ensure instant contact. Pye research development and quality control really do keep in touch with tomorrow.





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Pye 'Pocketione' Personal Radiotelephone New battery economy circuit Ea-tremely light-weight and compact -Reception free from noise and inter-ference - Minimum of controls -Transmit button automatically ea-tends antenna - Hearing ald socket -Eastly accessible batteries.

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Front Mounted Radiotelephone

Completely solid state - 5-8W R.F. output - 1-10 channels with solid state switching - Suitable for all climates - Meets all relevant speci-fications.

Pye 'Westminster'



Pye 'Bantam' Portable VHF Radiotelephone Fully transitionised transmittei and receiver Very high performagce receiver Crystal filter selectivity. 0-5W transmitter output 250mW audio power Long endurance with rechargeable or dry batteries Can be used with external antenna to give greater range Weatherprool.



Pye Single-Sideband Radiotelephone

Madiotelephone 125W (n.e.p.) R.F. output - Fully transistorised receiver - C.W. faci-liftes provided - Sideband selection by crystal filter - Carrier insertion for a.m. compatibility - Fixed or mobile application - Advanced transmitter dealon. applica design



Pye VHF Radiotelephone Fixed Station

Solid-state receiver and transmitter -10-15W R.F. output Field-effect transistors used in receiver Suitable for all climates - Electronic squelch -Designed to meet all relevant speci-tionates - Electronic squelch -



Pye 'Pioneer Radiotelephone Fully transistorised · For use with automatic. CB manual, or magneto exchanges · Weatherproof cabinet · Unattended operation over long perioda · Facility for fitting-privacy equipment · Optional single antenna operation.



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Pye UHF Radiotelephone Fixed Station Solid state receiver and transmitter -8-10W R.F. output . Very high R.F. selectivity using field-effect tran-sistors . Very low noise factor . Electronic soueich A.C. co r24V d.c. operation . Suitable for all climates Designed to meet all relevant speci-tications.

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Compact 3-circuit radio terminal -Fully -tensistorised channelling equipment - Frequency-shift signalling Continuous unattended operation in all parts of the world - Tweive standard plans for terminals and repeaters.

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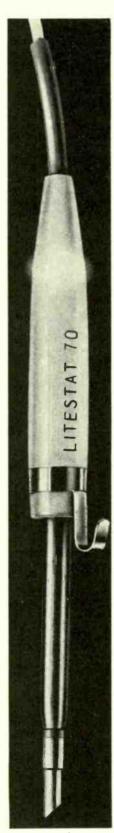
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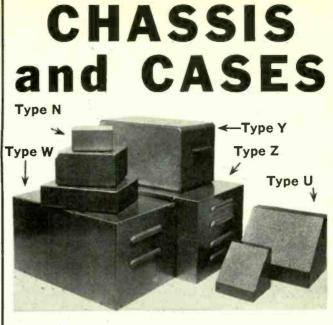
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N	4 × 4 × 2	11/-	Y 8×6×6	29/-
U	4 × 4 × 4		$Y 12 \times 7 \times 7 \dots$	45/-
U	$5\frac{1}{2} \times 4\frac{1}{2} \times 4\frac{1}{2} \dots$	17/-	$Y 13 \times 7 \times 9 \dots$	50/6
Ū	8 × 6 × 6	23/-	Y 15 × 9 × 7	53/6
Ŭ	$9\frac{1}{2} \times 7\frac{1}{2} \times 3\frac{1}{2} \dots$	e.	$Z = 17 \times 10 \times 9 \dots$ $Z = 19 \times 10 \times 84$	
Ŭ	15 × 9 × 9		$Z = 19 \times 10 \times 8\frac{1}{2} \dots$ *Height	78/-
w	8 × 6 × 6		Plus post and packi	ng.

Type N has a removable bottom, Type U removable bottom or back, Type W removable front, Type Y all-screwed construction, Type Z removable back and front.

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	TO F	TT C	DUR CASES		
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$4 \times 21 \times 11''$	6/-	$3 \times 2 \times 1^{\prime\prime}$	5/6
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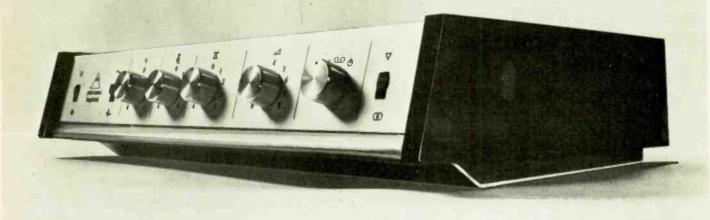
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The Peak Sound Englefield is a new system which assembles from laboratory designed modules to provide a cost-performance ratio which has never been bettered in high fidelity. Here is top-flight circuitry housed in a cabinet of elegantly original design which is both beautiful and completely practical back and front. By assembling these Peak Sound units, you can own one of the best high fidelity instruments you have ever heard or seen and all for a cost of about £38. The assembly is supplied complete down to the necessary connecting wires supplied colour coded, cut to length and stripped at the ends for soldering. You can use the Englefield Cabinet design to house either the 12 + 12 system as published in *Practical Wireless*, or the 25 + 25 watt system as approved for the *Hi-Fi News* Twin Twenty by Reg Williamson. Go to your stockist and ask to see and hear Peak Sound equipment now.

Matching F.M. Tuners will be available very shortly.

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THE SPECIFICATION

Using two Peak Sound PA.12-15's, driven simultaneously at 1 KHz from 240 V, mains supply.

Output per channel: 11 watts into 15Ω : 14 watts into 8Ω . (see spec. guarantee). Frequency bandwidth: 10Hz to 45 KHz for 1dB at

1 watt.

Total Harmonic Distortion at 1 KHz at 10 watt into 15. Ω —0.1%. Input sensitivities: Mag. PU.3.5 mV imp. R.I.A.A.

equalized Into 68 K Ω : Tape, 100 mV linear into 100 K Ω : Radio, 100 mV linear into 100 K Ω . Overload factor: 29 dB on all input channels.

Signal/noise ratio: -65 dB on all inputs. Vol. control max.

max. Controls: Volume. Treble. Bass. Low-pass Filter. Mono/Stereo: On/off: Balance. Using two PA.25-15 amplifiers, output is then 25

watts into 15 Ω or 8 Ω per channel at 1 KHz.

Power bandwidth for -1~dB at 20 watts R.M.S. Into 15 $\mathcal Q$ at less than 0.25% distortion is 20 Hz to 20 KHz.

THE MODULES

Englefield Amplifier Cabinet with			
front panel, knobs, sockets, cut and			
stripped wire, fuses, edge connec-			
tors. etc.	£6	0	0
Two PA. 12-15 power amp. built			
modules	£11	19	0
SCU/400 Pre-amp/Control module,			
built	£15	15	0
PS/45 Power Supply kit	£4	10	0
	£38	4	0
			-

Using two PA.25-15 modules at $\pounds11/15/-$ each and PS/68S Stabilized Power Supply Unit at $\pounds13/10/-$, total price for complete system comes to $\pounds58$ 15 0

		t. Jude's Rd., E			
Details of	Englefield sy	stems, please and	1	· · · · · · · · · · · · · · ·	 • • • • • • • •
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Some notes on Bridge Measurement by WAYNE KERR

Number 6 Radio-Frequency Bridges

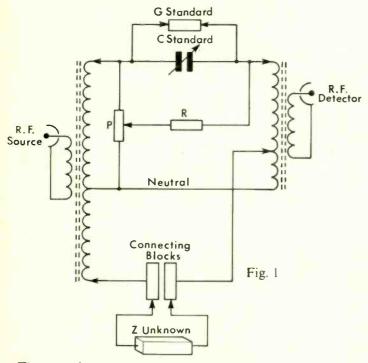
The first five issues in this series of notes have described some of the basic principles of low frequency bridges and also their application to the measurement of components using two, three and four terminal techniques.

Transformer Ratio Arm bridges can be designed to operate at radio frequencies up to about 250MHz where other forms of bridge based on transmission lines become practicable.

The design of a bridge required to operate at high frequencies demands careful attention to every aspect of the layout, and in particular to the series inductance introduced by connections between component parts of the bridge. Short lengths of conductor which are insignificant at low frequencies can resonate and introduce immense errors as the frequency is increased.

However, the neutral connection which is available from transformer ratio arms can be used to effectively cancel the series inductance of conductors in the following manner. If two strip connections are made to, say, a bridge standard, these are placed side by side and mounted above a plate connected to neutral. The loop current flowing in the strips will induce, in the plate, an equal and opposite current which cancels the magnetic field, thus reducing the loop inductance.

Figure 1 illustrates a practical circuit for a bridge capable of operating at frequencies up to 100MHz.



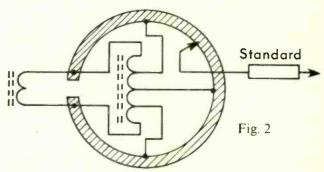
The transformers are formed by winding thin silver tapes on to ferrite or ferrous dust ring cores which are mounted inside individual screening cans.

> THE WAYNE KERR COMPANY LIMITED NEW MALDEN · SURREY · ENGLAND

The unknown impedance is connected to the blocks shown in the diagram which represent a shunt capacitance on the unknown side of the bridge. This capacitance is balanced by the standard variable capacitor and its value is so chosen that the capacitor is half engaged when the dials are at zero. An unknown reactance can therefore be balanced either by increasing the setting of the capacitor or decreasing it in the case of an inductive reactance. This feature is of particular value when transmission lines or aerial arrays are being evaluated.

Drums of low inductance resistors forming fixed conductance standards are arranged to engage with spring contacts. A variable conductance for interpolation is formed by means of a resistor R which is fed with a voltage derived from a resistive potential divider P.

Recently, a continuously variable potential divider has been developed which enables voltage division to be effected with great precison. This device is based on the magnetic field in a single turn loop and is illustrated in Figure 2.



The loop is connected to a winding which forms part of the left hand transformer shown in Figure 1. An autotransformer is connected across the loop and several taps are connected to give a predetermined voltage distribution round the loop. Separate loops can be used to drive resistive and reactive standards and one interesting feature of the arrangement is its ability to create a continuously variable inductance standard. In this case an air cored toroid can be employed whose external field is so small that the presence of metal objects near the coil has no measurable effect.

Radio-frequency bridge measurements require that considerable care should be taken in setting-up the apparatus. Any leakage of power from the source to the detector which by-passes the bridge network will give errors. Furthermore, if an aerial assembly is being measured, radiation from the aerial may be picked up by a badly screened detector and subsequently cancelled by a voltage of opposite phase in the operation of the bridge which will now balance at a false point on its scales. However, with a well screened detector and with soundly constructed connecting cables coupling the source and detector to the bridge, highly accurate measurements can be performed on both active and passive assemblies.

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I have to inform you that the arrangements at your Works at

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This approval covers the following:

Manufacture of electronic parts of assessed quality to the BS. 9000 series of specifications as follows:-

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The British Standards Institution Approval Number allocated to your Works

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and this should be quoted on all relevant documents.

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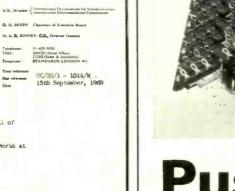
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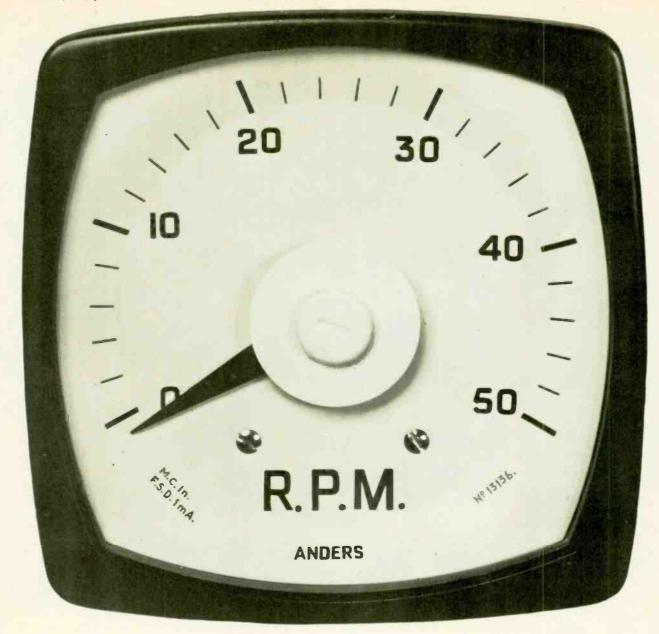
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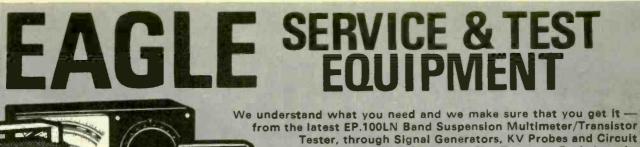
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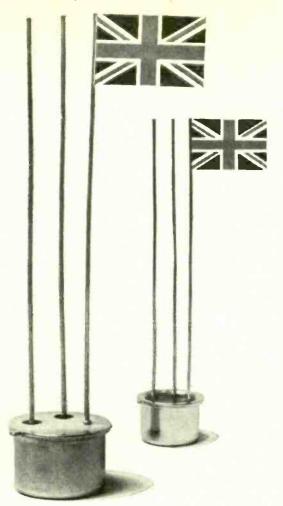
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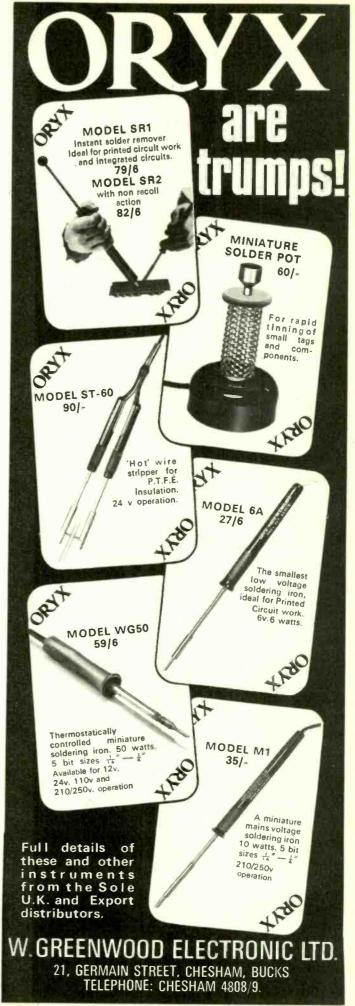
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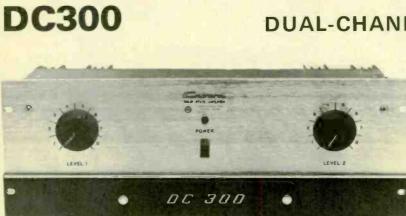
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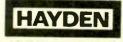
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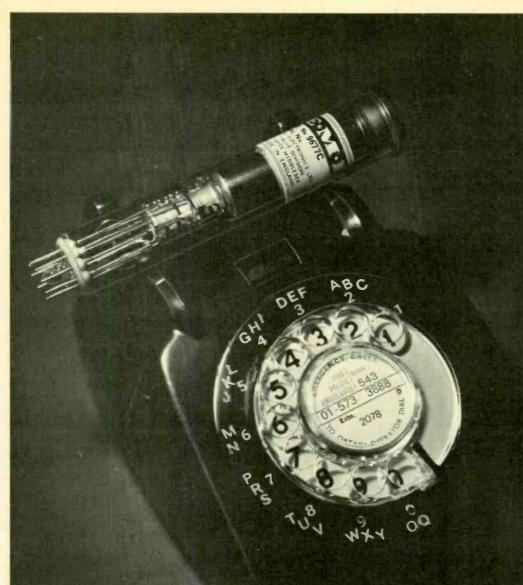


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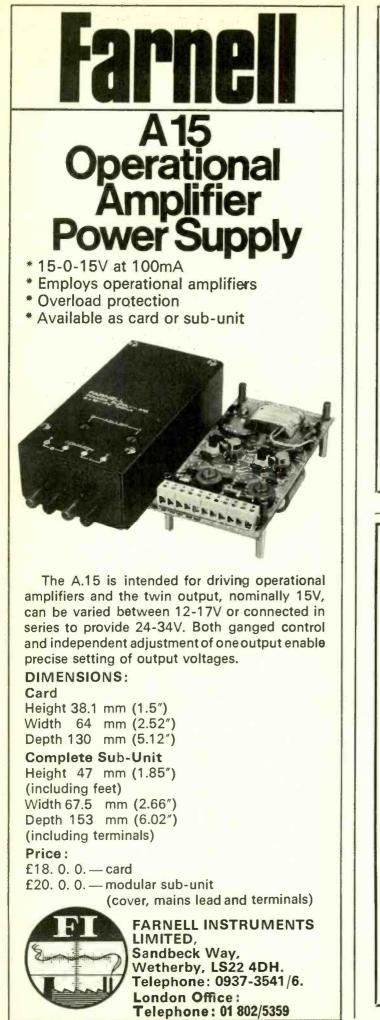
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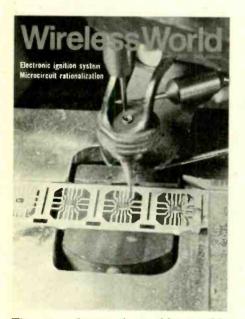
Wireless World

Electronics, Television, Radio, Audio

Fifty-ninth year of publication

January 1970

Volume 77 Number 1411



The macrophotograph on this month's cover shows wires being bonded on to integrated circuits at the Mullard Southampton works. On page 6 the future of linear i.cs is discussed.

OUR NEXT ISSUE

Loudspeaker performance: Paul Klipsch, originator of the Klipsch horn, compares horns and direct radiators

Ceramic pickups and transistor pre-amplifiers: are they incompatible? Matching: what is meant by this term?



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Contents

- 1 Editorial Comment
- 2 Capacitor-discharge Ignition System by R. M. Marston
- 6 The Future of Linear I.Cs by R. Hirst
- 8 News of the Month
- 10 Letters to the Editor
- 11 Circuit Ideas
- 13 Amorphous Semiconductors by J. E. Carroll
- 16 Application Notes
- 17 Magnetoresistance by B. E. Jones
- 19 Announcements
- 20 Low-distortion Bias and Erase Oscillator by D. Griffiths
- 23 Industrial Telemetry by R. E. Young
- 27 Active Filters 6 by F. E. J. Girling and E. F. Good
- 32 Personalities
- 33 Progress in Tape-recording Techniques by S. Feldman
- 34 World of Amateur Radio
- 35 Electronic Metronome by D. T. Smith
- 35 A Digital Christmas Tree
- 36 D.C. Bias in Push-pull Power Amplifiers by R. A. Smith
- 38 1970 U.K. Conferences & Exhibitions
- 39 New Products
- 44 Meetings
- 45 Test Your Knowledge questions and answers devised by L. Ibbotson
- 46 Literature Received
- 46 H.F. Predictions
- 48 Real & Imaginary by "Vector"
- A95 SITUATIONS VACANT
- A114 INDEX TO ADVERTISERS

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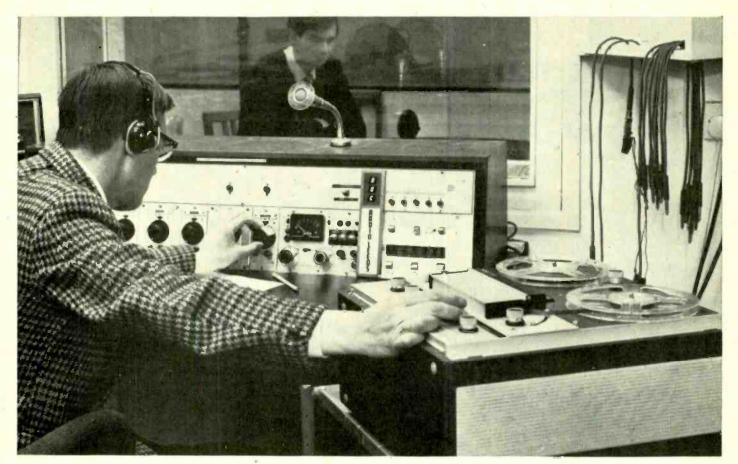
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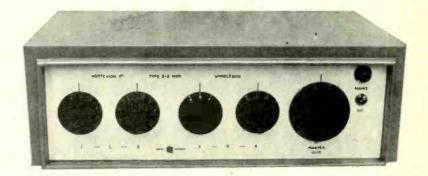


These electronic Stereo Mixers range from 2+2 to 5+5 input channels, with left and right outputs at 500 millivolts into 20K ohms up to infinity.

Separate control knobs are provided for L & R signals on each stereo channel so that a Mono/ Stereo changeover switch provided can give from four to ten channels for monaural operation, in which state the L & R outputs provide identical signals.

A single knob ganged Master Volume control is fitted, plus a pilot indicator.

The units are mains powered and have the same overall dimensions as monaural mixers.



STEREO MIXERS

Also available Monaural Electronic Mixers:-

4 Way Monaural Mixers
6 Way Monaural Mixers
8 Way Monaural Mixers
10 Way Monaural Mixers

3 Way Monaural Mixers with P.P.M.
4 Way Monaural Mixers with P.P.M.
6 Way Monaural Mixers with P.P.M.
8 Way Monaural Mixers with P.P.M.

50/70 WATT ALL SILICON AMPLIFIER WITH BUILT-IN 4 WAY MIXER USING F.E.T.'s. This is a high fidelity amplifier (0.3% intermodulation distortion) using the circuit of our 100% reliable 100 Watt Amplifier (no failures to date) with its elaborate protection against short and overload, etc. To this is allied our latest development of F.E.T. Mixer amplifier, again fully protected against overload and completely free from radio breakthrough. The mixer is arranged for $3-30/60 \Omega$ balanced line microphones, and a high impedance line or gram. input followed by bass and treble controls. Since the unit is completely free from the input rectification distortion of ordinary transistors, this unit gives that clean high quality that has tended to be lost with most solid state amplifiers. 100uV on 30/60 ohm mic. input. 100mV to 100 volts on gram/auxiliary input 100 K Ω .

CP50 AMPLIFIER. An all silicon transistor 50 watt amplifier for mains and 12 volt battery operation, charging its own battery and automatically going to battery if mains fail. Protected inputs, and overload and short circuit protected outputs for 8 ohms—15 ohms and 100 volt line. Bass and treble controls fitted. Models available with 1 gram and 2 low mic. inputs. 1 gram and 3 low mic. inputs or 4 low mic. inputs.

100 WATT ALL SILICON AMPLIFIER. A high quality amplifier with 8 ohms—15 ohms and 100 volt line output for A.C. Mains. Protection is given for short and open circuit output over driving and over temperature. Input 0.4v on 100K ohms.

200 WATT AMPLIFIER. Can deliver its full audio power at any frequency in the range of 30 c/s-20 Kc/s ± 1 db. Less than 0.2% distortion at 1 Kc/s. Can be used to drive mechanical devices for which power is over 120 watt on continuous sine wave. Input 1 m W 600 ohms. Output 100-120v or 200-240v. Additional matching transformers for other impedances are available.

30/50 WATT AMPLIFIER. With 4 mixed inputs, and bass and treble tone controls. Can deliver 50 watts of speech and music or over 30 watts on continuous sine wave. Main amplifier has a response of 30 c/s-20Kc/s \pm 1db. 0.15% distortion. Outputs 4, 7.5, 15 ohms and 100 volt line. Models are available with two, three or four mixed inputs for low impedance balanced line microphones, pick-up or guitar.

VORTEXION LIMITED, 257-263 The Broadway, Wimbledon, London, S.W.19

Telephone: 01-542 2814 & 01-542 6242/3/4

Telegrams: "Vortexion London S.W.19"

a53

www.americanradiohistorv.com



Scope for Going Places

The EM102 offers you a portable oscilloscope with an ideal performance at a realistic price. Just check its specification (10kV, 20nS/cm. writing speed plus sweep delay). It's designed for laboratory applications but fulfills the role of a completely self-contained unit for servicing purposes. Take it anywhere - it's mains or battery powered with a built-in battery option. Plug-in units are available with bandwidths from d,c, to 30MHz, voltage sensitivity down to 1mV/cm. If you have an application for an Oscilloscope for use in the laboratory, in the field, or in any unusual environment, write or ring today for information, details or an immediate demonstration.

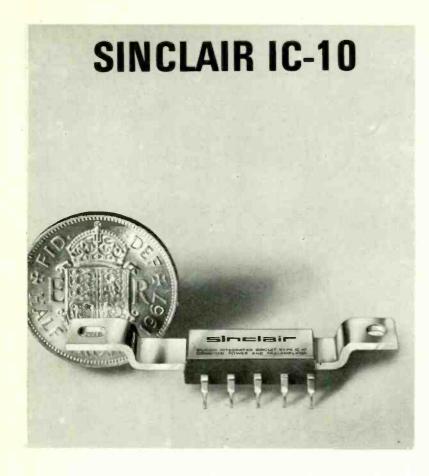
From £315.



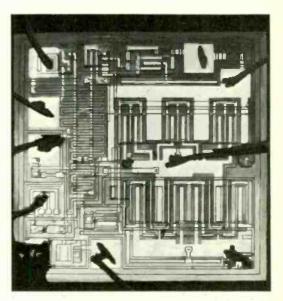
SE Laboratories (Engineering) Limited. North Feltham Trading Estate, Feltham, Middlesex. Telephone:01:890:1166&5246(sales):01:890:5876(works).Telegrams: Selab, Feltham. Telex: 23995. Northern Sales Office, Bessell Lane, Stapleford, Nottingham. Telephone: Sandiacre 3255.

WW-070 FOR FURTHER DETAILS

www.americanradiohistory.com



MONOLITHIC INTEGRATED CIRCUIT AMPLIFIER AND PRE-AMP



A 13 transistor circuit measuring only one twentieth of an inch square by one hundredth of an inch thick!

the world's most advanced high fidelity amplifier

The Sinclair IC-10 is the world's first monolithic integrated circuit high fidelity power amplifier and pre-amplifier. The circuit itself, a chip of silicon only a twentieth of an inch square by one hundredth of an inch thick, has 5 watts R.M.S. output (10w. peak). It contains 13 transistors (including two power types), 2 diodes, 1 zenor diode and 18 resistors, formed simultaneously in the silicon by a series of diffusions. The chip is encapsulated in a solid plastic package which holds the metal heat sink and connecting pins. This exciting device is not only more rugged and reliable than any previous amplifier, it also has considerable performance advantages. The most important are complete freedom from thermal runaway due to the close thermal coupling between the output transistors and the bias diodes and very low level of distortion.

The IC-10 is primarily intended as a full performance high fidelity power and pre-amplifier, for which application it only requires the addition of such components as tone and volume controls and a battery or mains power supply. However, it is so designed that it may be used simply in many other applications including car radios, electronic organs, servo amplifiers (it is d.c. coupled throughout), etc. Once proven, the circuits can be produced with complete uniformity which enables us to give a 5-year guarantee on each IC-10, knowing that every unit will work as perfectly as the original and do so for a lifetime.

MORE SINCLAIR DESIGNS ON PAGES FOLLOWING



SPECIFICATIONS

Output:	10 Watts peak, 5 Watts R.M.S. continuous
Frequency response	e: 5 Hz to 100 KHz ± 1dB
Total harmonic dis	tortion: Less than 1% at full output.
Load impedance:	3 to 15 ohms.
Power gain:	110dB (100,000.000.000 times) total.
Supply voltage:	8 to 18 volts.
Size:	1 x 0.4 x 0.2 inches.
Sensitivity:	5mV.
Input impedance:	Adjustable externally up to 2.5 M ohms.

CIRCUIT DESCRIPTION

The first three transistors are used in the pre-amp and the remaining 10 in the power amplifier. Class AB output is used with closely controlled quiescent current which is Independent of temperature. Generous negative feedback is used round both sections and the amplifier is completely free from cross-over distortion at all supply voltages, making battery operation eminently satisfactory.

APPLICATIONS

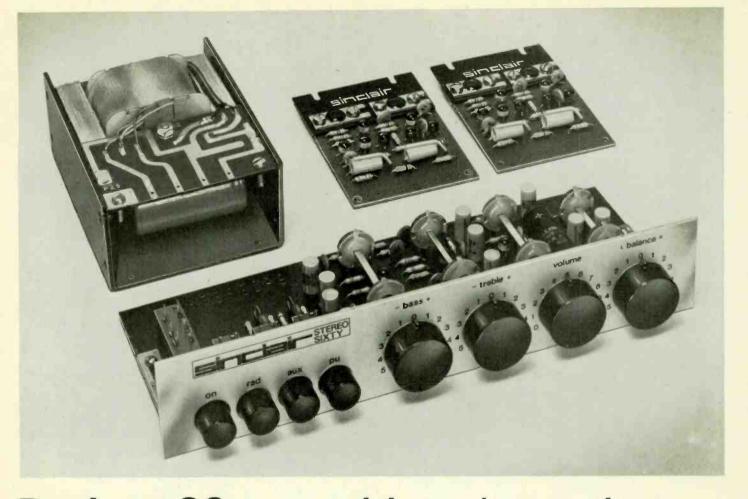
Each IC-10 is sold with a very comprehensive manual giving circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include stabilised power supplies, oscillators, etc. The pre-amp section can be used as an R.F. or I.F. amplifier without any additional transistors.





SINCLAIR RADIONICS LTD. 22 NEWMARKET ROAD, CAMBRIDGE

Telephone: 0223 52731



Project 60 an exciting alternative

The buyer of an amplifier today has a remarkably wide variety to choose from. It is unlikely that a purchaser would have real difficulty in finding a unit that met all his requirements, although the price might not be as low as could be wished. The only snags are that one's needs can change and that the technically correct amplifier may be physically inconvenient. If you are confident that there is an amplifier available, of the right size and price, which will meet all your needs for the forseeable future, then that is your best buy. If not, however, we can offer you another possibility which we believe to be an exciting alternative approach. That alternative is **Project 60**.

Project 60 is a range of modules which connect together simply to form a complete stereo amplifier with really excellent performance. So good, in fact, that only 2 or 3 amplifiers in the world can compare with it in overall performance.

The modules are: 1. The Z-30 high gain power amplifier, which is an immensely flexible unit in its own right. 2. The Stereo 60 preamplifier and control unit. 3. The PZ.5 and PZ.6 power supplies. A complete system comprises two Z-30's, one Stereo-60 and a PZ-5 or PZ-6. The power supplies differ in that the PZ-6 is stabilised whilst the PZ-5 is not. This means that the former should be used where the highest possible continuous sine wave rating is required. In a normal domestic application there will not be a significant difference between using either power unit unless loudspeakers of very low efficiency are being used.

All you need to assemble your system is a screwdriver and a soldering iron. No technical skill or knowledge whatsoever is required and, in the unlikely event of you hitting a problem, our customer service and advice department will put the matter right promptly and willingly.

Perhaps the greatest beauty of the system is that it is not only flexible now but will remain so in the future. We shall shortly be introducing additional modules which will include a comprehensive filter unit, a stereo F.M. tuner and an even more powerful amplifier for very large systems. These and all other modules we introduce will be compatible with those shown here and may be added to your system at any time.

Project 60 modules have been carefully designed to fit into virtually every known type of plinth or cabinet and templates provided enable you to position them. Only holes have to be drilled into the wood of the plinth and any slight slips here will be covered completely by the aluminium front panel of the Stereo 60. The Project 60 manual gives all the instructions you can possibly want clearly and concisely.



SINCLAIR RADIONICS LTD · 22 NEWMARKET ROAD · CAMBRIDGE Telephone: 0223 52731

TWENTY WATT R.M.S. (40 WATT PEAK) Z-30 **POWER AMPLIFIER**

The Z-30 is a complete power amplifier of very advanced design employing 9 silicon epitaxial planar transistors. Total harmonic distortion is incredibly low being only 0.02% at full output and all lower outputs. As far as we know, no other high fidelity amplifier made can match this specification, no matter what the price. Thus you can be utterly certain that your Project 60 system will do full justice to your other equipment however good it may be. The Z-30 is unique in that it will operate perfectly, without adjustment, from any power supply from 8 to 35 volts. It also has sufficient gain to operate directly from a crystal pickup. So in addition to its use in a high fidelity system you can use a Z-30 to advantage in your car or a battery operated gramophone for your children, for example. These, and many other applications of the Z-30, are covered in the Project 60 manual.

SPECIFICATIONS

Power output-15 watts R.M.S. (30 watts peak) into 8 ohms using a 35 volt supply; 20 watts R.M.S. (40 watts peak) into 3 ohms using a 30 volt supply.

Output-Class AB.

Frequency response: 30 to 300,000 Hz ± 1dB. Signal to noise ratio: better than 70dB unweighted. 0.02% total harmonic distortion at full output into 8 ohms and at all Distortion : lower output levels. Size : 3 ± x 2 ± x ± inches Input sensitivity: 250mV into 100 Kohms. Damping Factor: > 500 Loudspeaker impedances 3 to 15 ohms Power requirements: 8 to 35 V.d.c.

STEREO SIXTY

The Stereo 60 is a stereo preamplifier and control unit designed for the Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout and great attention has been paid to achieving a really high signal-to-noise ratio and excellent tracking between the two channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs. The tone controls are also very carefully designed and tested.

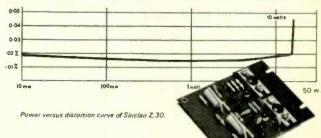
SPECIFICATIONS

Input sensitivities—Radio—up to 3mV: Magnetic Pickup-3mV Correct within ± 1dB on R.I.A.A. curve. Ceramic Pickup up to 3mV: Auxillary-up to 3mV.

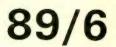
- Output—1 volt.
 Signal-to-holse ratio—better than 70dB.
- Channel matching—within 1 dB. Tone Controls—TREBLE + 15 to 15dB. 10 KHz: BASS + 15 to 15dB at
- 100 Hz.
- Power consumption 5mA
- Power requirement—PZ.5 or PZ.6.
 Finish—brushed aluminium front panel
- with black knobs.
- Mounting-on cabinet front by spindle bushes and adjustable brackets.

APPLICATIONS

High fidelity amplifier: car radio amplifier: record player fed direct from pick-up: intercom; electronic music and instruments; P.A., laboratory work, etc. Full details of these and many other applications are given in the manual supplied with your Z.30.



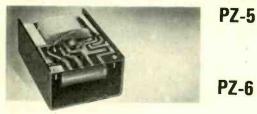
Z.30 Ready built, tested and guaranteed, with Z.30 manual.



PREAMPLIFIER AND CONTROL UNIT

Treble and bass control curves **STEREO SIXTY**

SINCLAIR POWER SUPPLY UNITS



30 volts unstabilised-sufficient to drive two Z-30's and Stereo 60 for the majority of domestic applications.

Price: £4. 19s. 6d.

35 volts stabilised-ideal for PZ-6 driving two Z-30's and a Stereo 60 when very low efficiency speakers are employed.

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Ready built, tested £9.19s.6d.

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GUARANTEE

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WW7013

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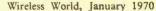
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WW-080 FOR FURTHER DETAILS

Wireless World, January 1970

WW-081 FOR FURTHER DETAILS





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Wireless World, January 1970



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Please note our address: SPECIALIST SWITCHES P.O. Box 3, CHARD, SOMERSET

Write for design charts and prices or TELEPHONE—CHARD 3439

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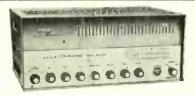


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*Many a product owes its success to the intelligent addition of an indicator light.

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PHOTOMULTIPLIERS. EMI 6097X at £8/10/- ea.

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DALE heat sink resistors, non-inductive 50 watt. Brand new, 15 ohms -6/6 ea.; 8.2K -5/-. Excellent dummy load

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TRANSISTORS BC 114; BC 116; Ex brand new equip-ment, Guaranteed perfect, Good lead length 2/- ex, Also RT Micrologic L 9914 I.C. 8 lead TO-5 can. Fan out 16 duai 2 input nand nor gate, capable of F/F action at 20 mc/s or non inverting gate or gate + invert. Good length leads. Guaranteed perfect 7/6 ca,

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SOLARTRON	7118.2 D.B. DC-9 mc/s £60
SOLARTRON	643 DC-15 mc/s £80
SOLARTRON	513/523 DC-10 mc/s £35
SOLARTRON	568 DC-6 mc/s £18
COSSOR	1035 DB. £20
COSSOR	1049; 1049 Mk. 3, DB. £22/10 and £30
	13A DB. £18/10/- ked and tested. Carriage 30/- extra.
	MARCONI
	tion Meter £45 Carr. 10/-

TF 956 Audio Frequency Wathmeter £15 Carr. 10/-TF 866 Magnification Meter £45 Carr. £1
 TF 762C UHF Generator £40 Carr. £1
 TF 995 AM/FM Generator 1.5-200 mc/s £120 Carr. £1
 TF 969 N. 5 Impedance Bridge £75 Carr. 30/-TF 144G Signal Generator, Serviceable, Clean £18 in Exceptional condition £30 Carr. 30/-TF 869 Willivolt neter £6 Carr. 10/-TF 105M Sine wave oscillator 0/40kc/s £14 Carr. £1
 TF 428B/1 Valve voltmeter £10 Carr. 10/-TF 338B Attenuator £5 Carr. 10/-TF 338B Attenuator £5 Carr. 10/-TF 701B Carrier Deviation Meter £4 Carr. 30/ SCH AETRON

TF 701B Carrier Deviation accord SOLARTRON Pulse generator POS 100C 50 c/#-1 mc/z £25 Carr. £1 Laboratory amplifier AWS51A. 15c/#-350kc/s £35 Carr. £1

Stabilised P.U. SRS 151A **£20** Carr. 30/-Stabilised P.U. SRS 152 **£15** Carr. 30/-Stabilised P.U. AS 516 & AS 517 **£3**, and **£6** Carr. 10/-

AVO Generator type TFN AM/FM 5-220mc/s £45 Carr. 15/-

Generator CT 368 2-225 mc/s **£50** Carr. 15/-Generator 50kc/s-80 mc/s **£16** Carr. 15/-Testmeter No. 1 **£14** Carr. 15/-Electronic Testmeter CT 38. Complete **£18** Carr. **£1**

Electronic Testimeter CT 38. Complete 116 Carr. 21 SFECIAL. Multimeter CT 2471A. Battery opera-ted, fully transistorised, sensitivity 100 M ohm/V, measures a.c./d,c. voltage (12m V-1200V scales, +/-3% / +/- 2% f.s.d.) a.c./d.c. current (12 mlcroA-1.2A scales, +/- 3% / +/- 2% f.s.d.) resistance (12 ohm-120M ohm scales, +/- 3% m.s.d.), h.f./ vhf/uhf, voltage with multiplier (4V-400V scales up to 50 MHz; 40 mV-4V up to 1000 MHz). Brand new (27 471B, as above by G. & E. Bradley 475 Carr. 30/-275 Carr. 30/-CT 471B, as above by G, & E. Bradley £75 Carr. 30/-

CINTEL Transistorised Nucleonic Scalers with adjustable discriminator. 6 meters display 0-9 giving count of 10 to the 5. In a new condition £30 Carr. 15/-Wide Range Capacitor Bridge 425 Carr. 15/-Sine and Pulse Generator type 1873 £25 Carr. 15/-

AIRMEC Valve Millivoltmeter type 264. 3MV-1V £20 Carr. £1 Counter type 85. 8 decades. Bright Vertical display gate facilities. Very good condition £30 Carr. 30/-Klystron Power Supply 698B £25 Carr. £1

SIGNAL GENERATOR/WAVEMETER type 61. 90 to 160 mc/s. Bullt in crystal markers. Standard mains input. Excellent condition £12/10/-. Carr. 30/-.

SIGNAL GENERATOR type CT 478 by W. H. Sanders. Freq. range 1.3 kmc/s to 4.2 kmc/s. Output power calibrated dbb. New £40. Carr. £2/10/-.

AM/FM OSCILLATOR Type 1 CT 212. 85 kc/s to 32 mc/s. As new £35. Carr. £1

HEINZ GUNTHER AM/FM Generator 9-200 mc/s. £30 Carr. £1

SIGNAL Generator CT 53. Complete with leads. Good condition. £10 Carr. 15/-

FREQUENCY Meter BC 221. £17/10/0; with built-in stab. P.U. £22/10/0 Carr. 15/- With Charts.

DIGITAL VOLTMETER type BIE 2114. 1 mV to 1kV DC. Auto decimal change. Excellent condition £65

MIC-O-VAC type 22 (CT54) Volta; Current; Ohmas, D Cto 200 mc/s with probe, leads etc. As new f8/10/0 P, & P. 10/- Plus-in MAINS PACK for Mic-ovac type CT 54, f2/10/0 P, & P. 6/-

VIBRATING REED ELECTROMETER type N 572 by ECKO. Range 10 to the -14. Max sensitivity FSD for by ECKO. Range 10 to the -14. Max sensiti 1 of 0.03 Micro-microamps. **£25** ea. Carr. £1

RACAL FREQUENCY COUNTERS SA 20 4 decades 100 kc/s £20 ea.

SA 20 4 decades 100 kc/s £20 ea. SA 21 6 decades 10 mc/s £45 ea.

CHILTMEAD

etc., at 'Chiltmead' prices. Callers welcome 9 a.m. to 10 p.m. any day.

SA 28 as SA 21 with convertor extending range to 30 mc/s $\pounds 65$ ea.

GRAUMONT-KALEE Flutter meter type 564 £35

FOR CALLERS. Always a large quantity of components, transformers, chokes, valves, capacitors, odd units,

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GENERAL RADIO Precision Capacitor type 722. 50 pf to 290 pf. 675 Model 222F at 455

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Miniature SCOPE type 1200B. Ideal TV servicing etc. Checked and tested (10 P. & P. £1 ADVANCE Signal Generator type D1. 2 mc/s to 190 mc/s. Sine and square mod. With original charts. Excel-lent condition £12/10/0 P. & P. £1

SERVOMEX. Stabilized D.C. Power supplies type DC3. 0.30 Volts. 0-7 amps. Separate voltage and current meters. £35. Type 38, bench mounting. 0-15 Volta. 0-2.5 amps. Separate voltage and current meters £25 TRANSIENCE AND ADDRESS OF A TRANSISTOR Stabilised Power Unit. 48v, 4 amp. Manufactured by E.M.I. Open chassis. Brand new. Highest quality. Size 10 j × 5 j × 6 j in. high. **£6** ea.

RACAL stabilizer unit 24V. raw DC in 20 volts 1 amp stab and 12V. 5 MA Zener stab out. Brand new condition. Size $5 \times 3 \times 6$ in. Complete with circuit diagram. 35/- ea. MAINS FREQUENCY METERS with large 270 deg. Dial calibrated 45 to 65 cycles 4in. diameter. In original service carton, £3/10/-. P. & P. 7/6.

19in. Rack Mounting CABINETS 6ft. high 2ft. deep. Side and rear doors. Fully tapped, complete with base and wheels. £12/10/0 Carriage at cost.

V.H.F. RECEIVER TYPE 715 by BCC. Complete, tested and working (Less crystal). 12v DC input. Ideal conversion 2 and 4 metres. In good condition. Supplied with conversion data. Only **63/10/0** P. & P. 7/6 ea.

MULLARD Transistorised Analogue to Digital Con-vertor Model I, 281. As new. **£20** Carr. 15/-

SUNVIC DC chopper Amplifier type DCA 1. Superb condition. £22/10/0 ea. Carr. 20/-

SEQUENCHAL TIMERS 8 individual timer circuits each with 0-100 sec calibrated dials. Ideal displays, processes, etc. Standard mains input **£25** Carr. 25/-

Ex- SOUTHERN ITV CAMERA with image orthicon tube and single lens. Mounted on tripod. Sold as seen. £40 ISOLATING TRANSFORMERS 240V in 240V 7 KVA out. As new. £25 ea. Carr. £2/10/-

DESK Telephones. Current type.'3 wire red, green, white. Ideal extensions etc. As new 43 ea.

MODERN Transistorised 10 line automatic desk top exchange. Complete with 10 A.E.I. Telephones, standard G.P.O. type dialibra to call any other line. Exchange size $16 \times 5 \times 6$ housed in attractive mahogany case. Complete set. **455**

METERS

ELLIOTT Portable Berfecting Voltmeter. 6" dial scaled 0 to 12.5 Ranges 12.5-25-37.5 Frequency range DC to 2.5 kc/s. Accuracy 0.5 fed. Adjustable feet, built in level, magnetic shield, movement lock. In fine quality wood case. **£20** P. & P. 15/-

EHT ELECTROSTATIC Ernest Turner etc. 0.750V £2 ea.; 0-3.5 KV £2/15/0 ea.; 0-5 KV £3/10/0 ea.; 0-7.5 KV £4/5/0 ea.

TAYLOR 100-0-100 Micro amp scaled size 4 \times 2" with internal lamp scaled 6-0-6 ~1/10/0 ea.

GRIFFIN & GEORGE 3" round in sloped open ended case with terminals 3 types 0/20; 0/100 and 0/250. All 50 c/s, £l ea.

TRANSFORMERS. All standard inputs. 18v 6 amp and 12v 1 amp. Sep. windings, 18/6 ea. 18v 12 amps at 43 ea. 18v 14 5mA. 4v 0.5 amp×2, 4v 1.1 amp. 43/10/- ea. 350-0-350 75mA, 5v 2 amps×2, 21/- ea. Gardners 6.3v 2A; 6.3v 1.5A; 6.3v 0.1A, Size $3 \times 1\frac{1}{2} \times 4\frac{3}{2}$ in. As new. 14/- ea.

Gardners, Potter. Multi 6.3's combine to give 48v at 4 amps or 6.3 at 45A. With 350-0-350 at 50mA.

4 amps (£2/10/- ea.

Earneko/Gardners. Potted. 475-60-0-60-475 at 160 mA; separate winding 215-0215 at 45mA; 6-3v 5A; 6-3v 0-75A; 5v 3A. As new. **£3** ea. Gardners 400-350-0-350-400 at 250MA; 0/4/6-3v 4 amp × 2; 0/4/6-3 2 amp; 0/4/5 3:5A. In original boxes. **£4/10/**, inc. post. Gardners 2kV 10MA. As new, **£3** incl. postage. Gardners 2kV 10MA and 4 volta×2. **£4/10/**- ea incl. postage

Parmeko 6.3v at 2 amp \times 4, 22/6 ea. Parmeko 65v 1 aup. Separate 0.18-24v at 0.5 amp. 30/- ea. Gard/Parm/Part. 450-400-0-400-450. 180 MA. 2 \times 6.3v. 43 ea.

E.H.T. Brand new 5kV 5MA with rectifier heater winding. Size $3 \times 3 \times 3$ in. 27/6 ea.

ADVANCE Constant Voltage Trans. 6 volts 50 watt. As new $\pounds 3$ P. & P. 10/-Gardners 4v 30 anp. Brand new $\pounds 1/10/0$ incl. postage. S for $\pounds 3/10/0$ incl. postage. Gardners 5v 30amp. Brand new $\pounds 2$ ea. incl. postage,

CHOKES, 5H: 10R; 15H: up to 120mA, 8/6 ea. Up to 250mA 12/6 ea. Large quantity LT. HT, EHT transformers. Your requirements, please. Panel switches DPDT ex eq. 2/6 ea.; DPST Brand new 3/6 ea.; DPDT twice, brand new 6/-; heavy duty DPST brand new 6/- ea.

PRECISION continually rotarable stud switches. Single pole. 80 way, can be stacked if required. £3 ea. PRECISION rotary stud switches 2 pole 12W size 2" sq., 1" shaft. £2/10/0 ea.

Min. SEALED 4 pole 3 way and 3 pole 4 way rotary switches, $\frac{1}{2}$ shaft $\frac{1}{2}$ dia. $\times \frac{1}{2}$ 10/- ea,

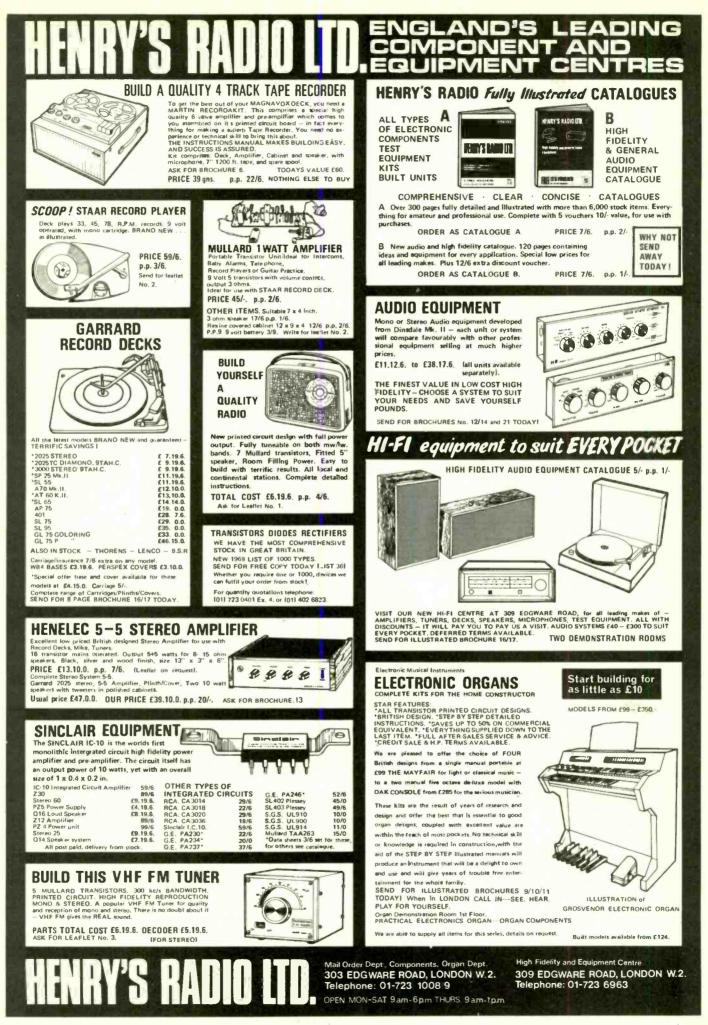
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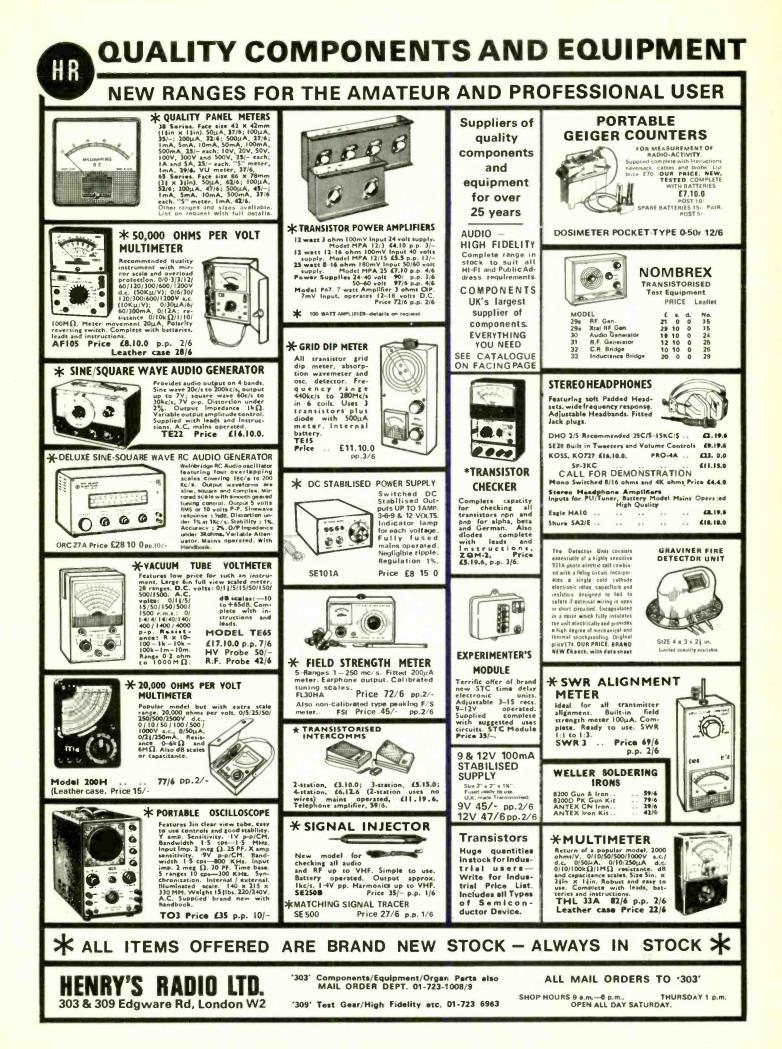
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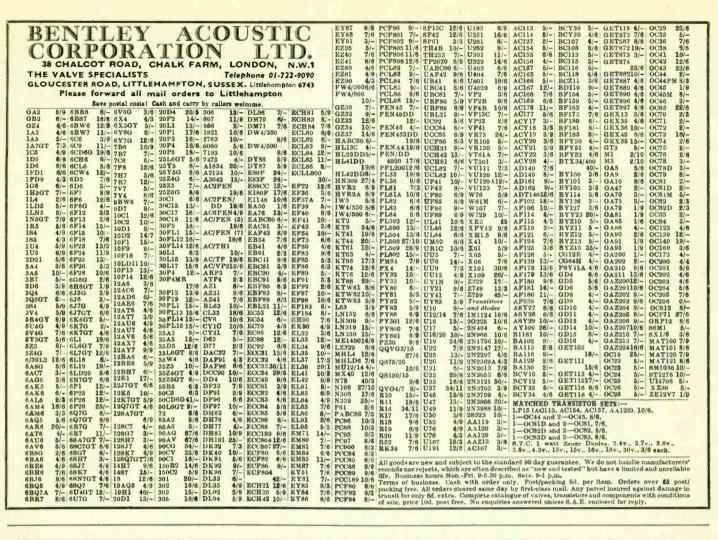












COPPER LAMINATE PRINTED CIRCUIT BOARD ($B_2^+ \times 5_3^+ \times 1_4^+$ in.), 2/6 sheet, 5 for 10/-. Also 11 × 9 in., 4/- ea., 3 for 10/-. HIGH SPEED MAGNETIC COUNTERS (4×1×1 in.) 4 digit. 6/12v. 24/48v. (state which), 6/6 ea. P.P. 1/-. ULTRASONIC BULK COMPONENT OFFERS CLEANERS 100 Capacitors (latest types) 50pF to .5µF. 250 Resistors 1 and 1 watt. 250 Resistors 1 and 1 watt. PYE OHMMETER TYPE 10B. 500v. test. .3 meg. ohm-20 k. meg. ohm. 200/250v. A.C. Brand new instrument £30. (Burndept B.E.352) 60 watt model. Supplied Brand New complete with stainless steel tank $9\frac{1}{2} \times 6\frac{1}{2} \times 4\frac{1}{2}$ in. £60. Carr, 20/-. 150 HI-Stab Resistors, 1, 1 and 1 watt. P.P. 30/. 25 Vitreous W/W Resistors, 5%. 12 Precision Resistors .1% (several standards POT CORES TYPE LA 3. 10/- ea. 2. FAST NEUTRON MONITORS (Burndept 1407C) for measuring neutrons in the energy range 0.15-15 meV. £100. 12 Precision Capacitors 1 and 2% (several standards Included). 12 Precision Capacitors 1 and 2% (several standards Included). 71 WAY PLUG & SOCKET (Painton Series 159) Gold plated contacts with hood & retaining clips. 30/- pair. 3. Radiation Monitors (Burndept BN 110 MK. V) 0-5/50/500/5k, c.p.s. Brand new. £100. Alpha and Beta Gamma probes available at extra cost. 50 WAY PLUG & SOCKET (U.C.L. miniature). Gold plated contacts 20/- pair. 34 way version 15/- pair. Electrolytics (miniature and standard sizes). ANY ITEM 12/6. ANY 5 ITEMS 50/-. 12 E VALVE MILLIVOLTMETER (Marconi TF899) 0-2v. complete with R.F. probe £8/10/- p.p. 10/-. PORTABLE RADIATION MONITORS (Burn-dept BN 132) 0-5/50/500/5k c.p.s. With built-in Gamma probe. Brand new. £50 complete with carrying harness. LOGIC BOARDS with 31 ACY40s-38 dlodes etc 20/- e8. TELEPHONE DIALS (New) 20/- ea. CO-AX RELAYS (magnetic devices) 1 change-over 12 v.w 20/- ea. RELAYS (G.P.O. '3000'). All types. Brand new from 7/6 each. 10 up quotations only. S.A.E. for literature, 10% discount for **Educational Authorities** EXTENSION TELEPHONE (Type 706) Black or 2 tone Grey. 65/-. P.P. 5/-. ELECTRONIC ORGAN BUILDERS. We now have in stock P.C. boards built to computer standards. Each board is a complete 4 octave divider (41 × 3 in.). All connection data UNISELECTORS (Brand new) 25-way 75 ohm. 8 bank ½ wipe 65/-. 10 bank ½ wipe 75/-. SPEAKER SYSTEM (20×10×10 in.). Made to spec. from # in, board, Finished in black leathercloth. 13×8 in. supplied. 30/- each. Set of 13 (gives 5 octaves to keyboard) from ‡ In, board. Finished in black restriction, to com-speaker with twin tweeters complete with cross-over 50c/s-20k/c, £7.10, P.P. 10/-. SPEAKER CABINET KIT, Above mentioned cabinet only. In kit form which you may assemble and Cover to your own

40/- PP 5/-

SPEAKER BARGAINS. E.M.I. 13×8 in. with double Tweeters 15 ohm, 65/-, P.P. 5/-. As above less tweeters 3 or 15 ohm, 45/- ea., P.P. 5/-.



CYLINDRICAL FANS (Solartron). Overall size 16 × 5²/₄ × 3¹/₂ in., air outlet 12 × 1¹/₂ in 240v A.C. 50/- each. P.P. 7/6 (New).

LEVEL METERS (1 + × + in.). 200 micro-amp. Made in Germany. 15/- each. PHOTOMULTIPLIERS 6262 and 6262b. £15 ea.

RELAYS H.D. 2 pole 3 way 10 amp. contacts. 12v.w. 7/6 ea. LIGHTWEIGHT RELAYS (with dust-proof covers) 4 c/o contacts. 24v. 500 ohm 7/6 ea. DIODE LOGIC BOARDS contains 10 diode gating circuits which convert any one of 10 inputs into an equivalent binary code, 10/- each.

TRANSFORMERS

E.H.T. TRANSFORMER 2100-0-2100v. 40m/a. 75/-P.P. 10/-E.M.T. TRANSFORMER (Parmeko 'Neptune') 3,000v. 280 m.a. £12/10/0. P.P. 50/-L.T. TRANSFORMER 60v. 8 amp. £5. P.P. 15/-, L.T. TRANSFORMERS Prim. 200/250v. Sec. 0-1/0-3/0-9/0-27v. 30 amp. £7.10. 15 amp. £5. P.P. 15/-, L.T. TRANSFORMERS Prim. 200/250v. Sec. 0/25/35v. 30 amp. £7.10. P.P. 20/-, STEP-DOWN TRANSFORMERS Prim. 200/250v. Sec. 115v. 1.25 amps, 25/- ea. P.P. 5/-,

115v. 1.25 amps, 25/- ea. P.P. 5/-. L.T. TRANSFORMERS Prim. 240v. Sec. 8/12/20/25v. 3.5 amp models 20/-; 5 amp model 25/-. P.P. 5/6.

L.T. TRANSFORMERS Prim. 240v. Sec 14v. 1 amp 10/-

ELECTRIC SLOTMETERS (1/-) 25 amp. L.R. 240v. A.C. QUARTERLY ELECTRIC CHECK METERS, 40 amp 240v. A.C., 20/- ea. P.P. 5/-.

REED RELAYS 4 make 9/12v. (1,000 ohm.) 12/6 ea. 2 make 7/6 ea. 1 make 5/- ea. Reed Switches (12 in.) 2/-£1 per doz

ea. Liper doz. SUB-MINIATURE REED RELAYS (1in.×±in.). Weight ≟ oz. Type 1. 960 ohm. 3/9v. 1 make. 12/6 ea. Type 2. 1800 ohm. 3/12v. 1 make. 15/- ea.

PRECISION CAPACITANCE JIGS. Beautifully made

RECISION CAPACITANCE JIGS. Beautifully made with Moore & Wright Micrometer Gauge. Type 1, 18,5pf-1220pl. £10 ea. Type 2. 9.5pf-11,5pf. £6 ea. TC CRYSTAL LOCKED OSCILLATOR (Synthesiser). 1K/c-20M/c. Output 0 dbm. 80 db att. in 1 db steps. Precision crystal oven. Locks oscillator at each 100K/c. Separate locked oscillator from 0-100K/c. £150 in excellent condition. condition



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Wireless World, January 1970



21kW FAN HEATER

Three position switching to suit changes in the weather. Switch up for full heater (3kW), switch down for half heat (1kW), switch central blows cold for summer cooling-allustable thermostat acts as auto control and safety cut-out. Complete kits [23,15,0. Post and Ins. 7/6. Made-up model £4,5,0. Post and Ins. 7/6. Made-up ins. 7/6.

FLUORESCENT CONTROL KITS

FLUORESCENT CONTROL KITS Each kit comprises seven liems—Choke, 2 tube ends, starter, atter holder and 2 tube ellps, with wiring instructions. Suitable for normal fluorescent tubes of the new "Grout" tubes for fish tanks and Indoor plants. Chokes are super-silent, mostly resin filled. Kit A-15-20 v. 1946. Kit B-30-40 w. 1946. Kit C-80 v. 23.46 Kit E-65 w. 1946. Kit MFI is for film. 9 in and 121m. miniature tubes, 1946. Dif C-80 w. 23.46 Kit S-05 w. 1946. Kit MFI is for film. 9 in and 4/6 for one or two kits then 4/6 for each two kits ordered. Kits C, D and E 4/6 on first kit then 3/6 on each two kits ordered.

BLANKET SWITCH

Double pole with neon let into side so luminous in dark, ideal for dark room light or for use with waterproof element-new plastic case, 5/6 each. 3 heat model 7/6-

BLANKET SIMMERSTAT

Although looking like, and fitted as, an ordinary blanket switch, this is in fact a device for switching the blanket on for varying time periods, thus giving a complete control from off to full heat. Also suitable for controlling the temperature of any other appliances using up to 1 ann. Listed at 27/6 each, we offer these while our stocks last at only 12/8 each.

REED SWITCHES

Giass encased, switches operated by external magnet-gold weided contacts. We can now offer 3 types: Ministure: In. long x-approximately jin. diameter. Will make and break up to jA up to 300 volts. Price 2/8 each.

make and break up to jA up to 300 volts. Price 2/8 each. **Standard**, 21n long × 3/16in. diameter. This will break currents of up to 1A, voltages up to 250 volts. Price 2/- each. **18**/- per dozen. **Flat**. Plat type, 21n. long, 1ust over 1/16in. thick. approxi-mately jin. wide. The Standard Type flattened out, so that it can be fitted into a sumailer space or a larger quantity may be packed into a square solenoid. Rating 1 am 200 volts. Price 6/- each. £3 per dozen. Small ceramic magnets to operate these reed switches **1/3** each. **12**/- dozen.

HIGH CAPACITY ELECTROLYTICS

Brad new, not ex-equipment. 100 mtd. 25v., 1/3 each 12/- doz. 200 mtd. 25v., 1/6 each 15/- doz. 500 mtd. 12v., 16 each 15/- doz. 500 mtd. 12v., 3/- each £1.10.0 doz. 500 mtd. 12v., 4/9 each £3.0.0 doz. 10,000 mtd. 15v., 8/9 each £3.0.0 doz. 10,000 mtd. 15v., 8/9 each £3.0.0 doz. 15,000 mtd. 10v., 10/6 each £5.0.0 doz. 15,000 mtd. 13v., 20/9 each £3.0.0 doz. 15,000 mtd. 13v., 20/9 each £3.0.0 doz. 15,000 mtd. 13v., 20/9 each £3.0.0 doz. 15,000 mtd. 13v., 40/- each £3.0.0 doz.

TELESCOPIC AERIAL

For portable, car radio or transmitter. Chrome-plated-six sections extends from 71 to 47in. Hole in bottom for 6BA screw. 7/6.

TOGGLE SWITCH with fixing ring. 1/6 3 amp 250v. each 15/- doz. C

80 OHM BALANCED ARMATURE EAR PIECE Usable as microphone or loudspeaker, 4/6 each.

MINIATURE EAR PIECE As used with imported pocket radios. 1/6 each 15/- doz.

ISOLATION SWITCH

20 Amp D.P. 250 Volts. Ideal to control Water Heater or any other appliance. Neon Indicator shows when current is on, 4/6 48/- per dozen.

FLEX BARGAINS

FLEX BARGAINS Screened 3 Core Flex. Each core 14/0076 Copper PVC Insulated and coloured, the 3 cores laid together and metal braided overall. Frice £3.15.0 per 100 yds. coll 15A 3 Core Non-Mitk Flex. 70/0076 Insulated coloured cores. protected by tough rubber sheath, then black cotton braided with white threat. A normal domestic flex as fitted to 8kW fires. Regular price 3/6 per yd. 30 yd. coll £4, or cut to your length 2/. per yard. 10A 3 Core Non-kink Flex. A shove but cores are 28/0076 Copper. Normai price 3/6 per yd. 100 yd. coll £7.10.0, or cut to your length 1/9 yd. 6A 2 Core Flex. As above, but 2 cores each 23/0076 ns used for Vacuum Cleaners, Electric Blankets, etc. 39/6100 yd. coll.

Polythene insulated 12-way strip. 2/6 each 24/- doz. 15/20 AMP CONNECTORS

13 AMP FUSED SWITCH - 14

Made hy G.E.C. For connecting water heater etc., into 13 amp ring main. Flush type 3/6 each 30/- doz. Metal boxes for surface mounting 1/6 each 15/- doz.

2 Da MICRC SWITCH changeover contacts. 1/9 each amp.

18/- doz. SUPPRESSOR CONDENSER TCC

1 mfd. 250v. A.C. working metal cased with fiting lug. 1/9 each 18/- doz. HEAT & LIGHT LAMP 275W, internally mirrored bulb, with bc. end for plugging into lamp holder. 19/6 each plus 4/6 post and insurance.

Into into noider. 19/6 each plus 4/6 post and insurance. TUBULAR TYPE BY PHILLIPS 500W. 29/6 plus 4/6 post and insurance. 750 MICRO AMP MOVING COIL METER 2/in. flush mounting, es-w.D. 15/- each plus 3/6 post and Insurance for any quantity.

DIGITAL CLOCK

The second second

An imposing instrument ideal for modern reception centre or for Managing Director's office-definitely a showpice to create interest and efficiency-mains frequency controlled as always keeps right time without adjustment-in black semi-matt perpex case-made up, tested and guaranteed-offered for only the cost components £29.10.0, post and insurunce 10/-.



100

EN I

SOLDER GUN

A must for every busy man, gives almost instant heat; also illuminates job. 100 watt 220/240v. 39/6 (saves you over 30/-), post and ins. 4/6. B1G JOB 250 watt model 99/6 (saves you over £3.10), post and ins. 6/6.

MINIATURE EXTRACTOR FAN

Beautifully made by famous German Company. PAPST System, 230/240 A.C. Mains operated, size 3 ln. x 3 ln. x 2 ln. Made for instrument cooling but kleal to incorporate in a cooker hood, etc. 65/= P, & p. 2/9.

HORSTMANN 'TIME & SET' SWITCH

(A 30 Amp Switch), Just the thing if you want to come home to a warm house without it costing you as fortune. You can delay the switch on time of your electric fires, etc., up to 14 hours from setting time or you can use the switch to give a boost on period of up to 3 hours. Equally suitable to control processing. Regular price probably around 25. Special snip price 29/8. Post and ins. 4/8.

DISTRIBUTION PANELS

D

Just what you need for work bench or lab. 4×13 smp sockets in metal box to take standard 13 amp fused plugs. Bupplied complete with 6 feet of heavy cable and 13 amp plug. Bimliar panels advertised at 55. Our price: Kit of parts **39**/**6**, plus 3/6 post and insurance. Made up **45**/- plus 4/6 P. \pm 1.

24 HOUR TIME SWITCH

Maina operated. Adjustable Contacts give on/off per 24 hours. Con-tacts rated 20 amps. repeating mechanism so ideal for shop window control, or to switch hall light (andt-burgiar precaution) while you are on holiday. Made by the famous Smiths Company. This month only 39/8 complete with perspec over, new and unused, plus 3/6 postage and insurance, a real snip which should not be missed. ELECTRIC CLOCK WITH As AMP SWITCH Made by Smith's, these units are as fited to many top quality cookers to control the orese. The clock is mains driven and frequency controlled so it is extremely accurate. The two small dula enable switch on and off times to accurately set. Ideal for switching on tape recorders. Offered at only a less than the value of the clock alone—post and insurance 2/9.

DOUBLE ENDED MAINS MOTOR

On feet with holes for screw-down fixing. To drive models, oven, blower heater, etc. 10/- each, plus 3/6 post and insurance. 6 or more post free.

DIAMOND H OVEN THERMOSTAT

Type 20 TH with capililary tube and sensor, 20 amp A.C. type as fitted to many cookers adjustable by control knob (not supplied). 12/6 each.

BATTERY OPERATED TAPE DECK

With Capstan control. This unit is extremely well made and newsures approx. 6 \times 3 \times 21n, deep. Has three pisnos key type controls for Record, Playback and Rewind. Motor is a special heavy duty type intended for operation off 4/3 rolts. Supplied complete with 2 spools ready to install. Record Replayhead is the sensitive M4 type intended for use with transistor, amplifier. Price 79/6. Post and insurance 4/6.

PROTECT DEVICES

DEVICES Fransistors, etc., which use heat sinks can easily be pro-tected; simply make the contact thermostat part of the heat sink. Notors and equipment generally can also be adequately protected by having thermostats in strategic apots on the casing. Our contact thermostat has a cali-brated dial for setting between 90° 100°P. or with a dial removed range setting is between 90° to 800°P. Price 10/-s.

ATLAS SLIMLINE FLUORESCENTS THE TWENTYLITE AND TWIN FORTY A Fluorescent lighting



HI FI BARGAIN FULL FI 12 INCH LOUDSPEAKER. This is undoubtedly one of the finest loudspeakers that we have ever offered, produced by one of the country's most famous makers. It has a fle-cast metal frame and is strongly recommended for HI-FI load and Rhythm Guitar and public address. Flux Densky 11.000 gauss—Total Flux 44,000 Maxwells—Power Handling 15 watts R.M.B.—Come Moulded fibre—Freq. response So-10.000 c.p.s.—specify 3 or 15 ohms—Main resonance 60 c.p.s. —Chasis Diam. 12in.—12i over mounting lugs—Baffle hole 11in, Jaim.—Mounting holes 4, holes—in. diam. on pitch circle, 11 jin. diam.—Overall height 5 jin. A 56 speaker offered for only 53.9.6 plus 7/6 p. 4; p. Don't miss this offer. 15in. 30 watt 57.19.8. 18in. 100 watt 5224.10.0.

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Will dim incandescent lighting up to 600 wait from full brilliance to out. Fitted on M.K. flush plate, same size and fixing as standard wall switch so may be fitted in place of this, or mount on surface. Price complete in heavy plastic box with control knob $\pounds3.19.6$.





Winter is coming but act today and you won't dismay. This heater unit is the very latest type, most efficient, and quiet running. Is as fitted in Hoover and blower heater costing £15 and more. We have a few only. Comprises motor, impeller, 2kW. element and lkW. element allowing switching 1,2 and 3kW, and with thermal safety cut-out. Can be fitted into any metal line case or cabinet. Only meed on/off awitch. **79/6**. Postage and insurance 6/6. Don't miss this. ELECTRONICS (CROYDON) LTD

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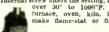
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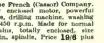
THERMOSTATSThe "A" is amp, for controlling room heaters, green-foures, etting cupboard. Has spindle for pointer knobs, built of the spindle state of the spin state of the spin state of the spindle state of the spin state of the spin state of the spindle state of the spin state of the spin state of the spindle state of the spin state of the spin state state of the spindle state of the spin state of the spin state of the spin state state of the spin state state of the spin state state state state in and out at a spin state state state state state the spin state state of the spin state state state state in and out at a spin state state state state state the spin state state of the spin state state state state in and out at a spin state state state state state the spin state state state state state state state state in a spin state state state state state state state the spin state state state state state state state in a spin state state state state state state state state the spin state state state state state state state in a spin state state state state state state state in a spin state state state state state state state state in a spin state state state state state state state state in the spin state state state state state state state state in the spin state state state state state state state state in state sta





THERMOSTATS





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MINIATURE WAFER SWITCHES

Male by famous Smiths Company. Very powerful, although only quite small. Overail dimensions approx. 14in. deep by 2in. dia. Available with the following appeed. Revs. per dour 1, 2, 4, 6, 12, 20, 30. Revs. per dour 1, 2, 4, 6, 15, 30, 60. 17/6 each.

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Precision made—as used in recei-decks and tape recorders—ideal at for extractor fans. blower, heater New and perfect. Snip at 9/6. Posto 3/- for first one then 1/- for eact ordered. 12 and over post free.

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10, 12, 15, 18, 22, 2		10 off	25 off	100 off
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SKELETON PRE-SET POTENTIOMETERS (Carbon) High quality pre-sets suitable for printed circuit boards of 0·1in. P.C.M. 100 ohms to 5 Megohms (Linear only). Miniature: 0·3W at 70°C. ±20% below 4M. ±30% above 4M. Horizontal (0·7in +0·4in. P.C.M.) or Vertical (0·4in. × 0·2in. P.C.M.). Subminiature: 0·1W at 70°C. ±20% below 2·5M. ±30% above.

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ubmini	iature	(all)	alues in a	LF)	iunaru.) —	10/0 00 10	0 70.	
V				32	64	125	250	400
-4V			6-4	25	50	100	200	320
.0V			4	16	32	64	125	200
6V			2.5	10	20	40	80	123
25 V			1.6	6:4	12.5	25	50	80
OV			1	4	8	16	32	50
4V			0.64	2.5	5	10	20	32
rice			1/4	1/3	1/2	1/-	1/1	1/2
mall (a		es in		1/3	•/ ▲	U=	1/1	4/4
V			800		1.250	2.00	0	3,200
·4V			640		1.000	1.60		2.500
ov			400		640	1.00		1.600
6V			250		400	64		1.00
5V			160		250	40		64
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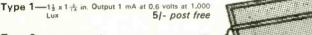


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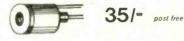
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When mounted on an aluminium heat sink 1 in, $x \frac{1}{2}$ in, $x \frac{1}{2}$ in. Supplied complete with suitable lenses, full Technical Data and Application Sheets, including Line of Sight Speech Link.



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Basic fibre optic components that demonstrate new ways of employing light in serious appli-

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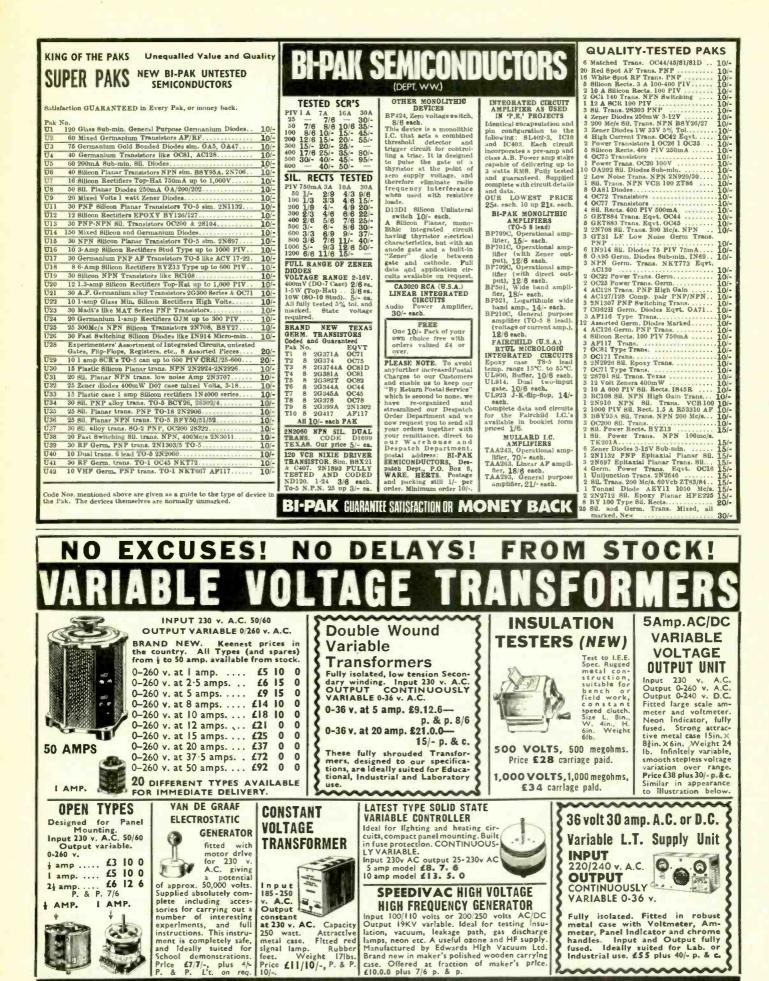
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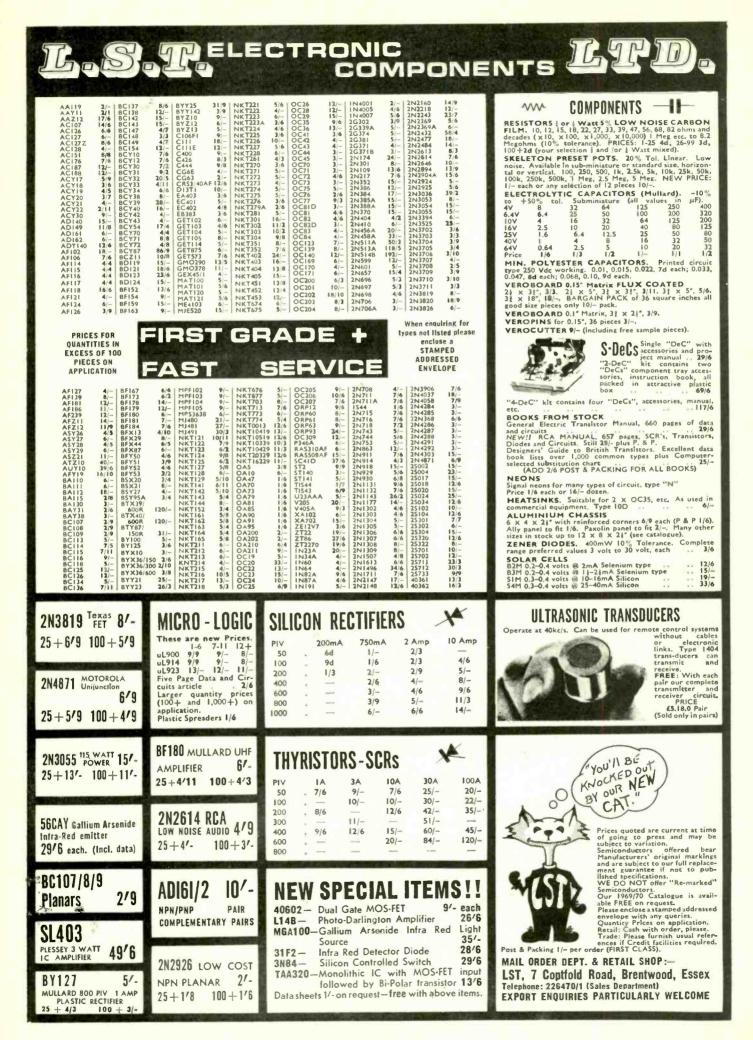
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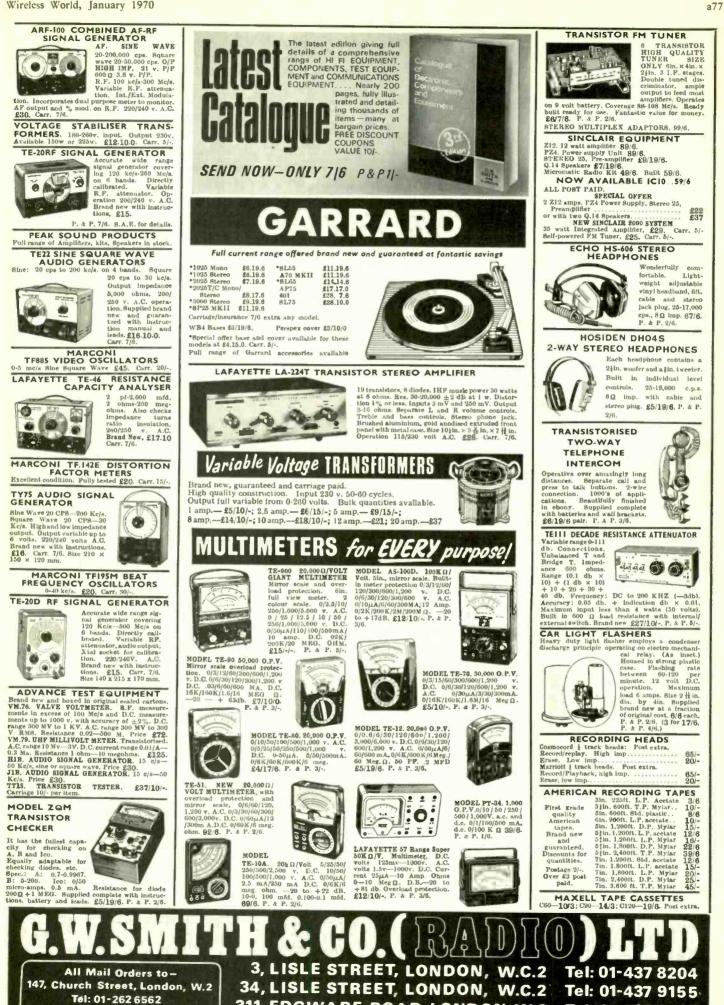


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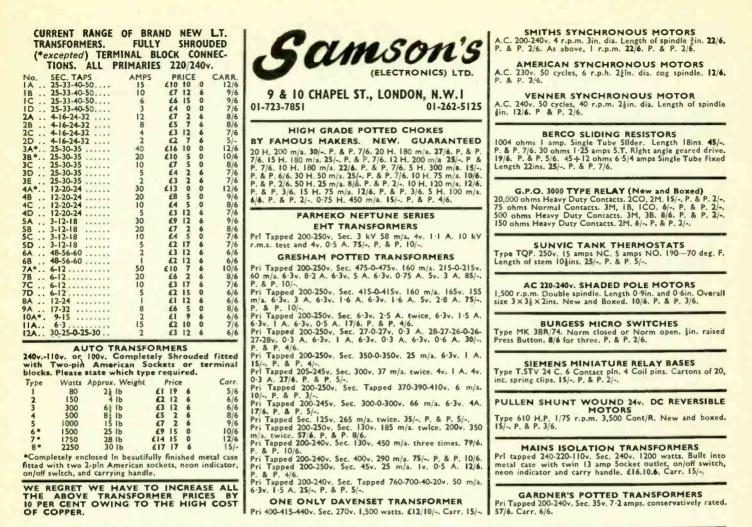
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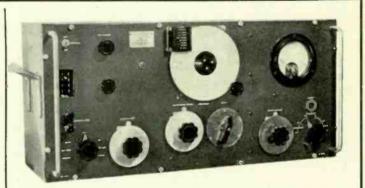
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ck the vi 13 o 22 o 36 o 91 o 220 o 470 o or ou	alues required. hms 560 hms 750 6 hms 1 k c hms 1.5 k hms 2.2 k hms 2.4 k r selection (mixed) 6 MICA/CERAMIC n following values. 5 pf 12 6 of 15	ohms ohms ohm ohm ohm ohm ohm B/6d. per 10 POLYST Tick those pf pf pf	3.3 k 3.6 k 4.3 k 5.6 k 5.6 k 6.8 k 7.5 k	c ohm c ohm c ohm c ohm c ohm c ohm c ohm	10 k ohm 16 k ohm 18 k ohm 22 k ohm 24 k ohm 27 k ohm 30 k ohm	39 43 47 51 62 75 82	k ohm 9 k ohm 13 k ohm 36 k ohm 43 k ohm 47 k ohm 56 k ohm 62 Total: Total: 180 pf 2 190 pf 3 200 pf 4	1 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 3/- per doz 50 pf 30 pf 20 pf	1.5 1.8 3.6 5.1 6.2 7.5	meg ohm meg ohm meg ohm meg ohm meg ohm meg ohm	9.1 meg oh
ck the vi 13 o o 22 o 36 o 47 o 220 o 47 o 0 o 470 o 0 o 7 ou: 220 o 470 o 0 o 7 ou: 200 o 200 o 2	alues required. hms 560 hms 750 6 hms 1 k c hms 1.5 k hms 2.2 k hms 2.4 k r selection (mixed) 6 MICA/CERAMIC n following values. 5 pf 12 6 of 15	ohms ohms ohm ohm ohm ohm ohm ohm ohm ohm ohm ohm	3.3 k 3.6 k 4.7 k 5.6 k 6.8 k 7.5 k 00. YRENE 25 pf 27 pf 30 pf 39 pf	cohm cohm cohm cohm cohm cohm cohm cohm	10 k ohm 16 k ohm 18 k ohm 22 k ohm 24 k ohm 24 k ohm 30 k ohm 30 k ohm 30 k ohm 30 pf 82 pf 100 pf 125 pf	39 43 47 51 62 75 82 75 82 135 pf 135 pf 138 pf 170 pf IESS GANC SKEL VOLU TELE 1 THIN CO-A:	k ohm 9 k ohm 13 k ohm 36 k ohm 43 k ohm 47 k ohm 56 k ohm 62 Total: Total: 180 pf 2 190 pf 3 200 pf 4	1 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 3/- per doz 50 pf 130 pf 130 pf 130 pf 130 pf 130 pf 130 pf 140 150 cf 150 cf	1.5 1.8 3.6 5.1 6.2 7.5 cn of any 580 pf 300 pf 300 pf 300 pf 300 pf 200 pf Total: 2/6d. each - dozen. neg. with OLS. Phildensers, yds 1/-, 1	meg ohm meg ohm meg ohm meg ohm meg ohm meg ohm one value. 1,000 pf 1,100 pf 1,500 pf 2,200 pf 2,200 pf u. b. D.P. switch. : ilips. Contain 10/- each. (CC 00 yds 7/6d.)	2,700 pf 3,000 pf 6,200 pf 2/- each. 11' 7-way ca sst £3/3/)
ck the vi 130 oc 220 o 36 o 470 o 91 o 220 o 470 o o 470 o 0 470 o 0 9 4 9 10 4 200 o 13.9 pf 300 o 1200 o 120	alues required. hms 560 hms 750 hms 1.5 k hms 1.5 k hms 1.8 k hms 2.2 k r selection (mixed) 6 MICA/CERAMIC, n following values. 5 pf 12 6 pf 15 8 pf 18 f 10 pf 22 20 20 20 20 20 20 20 20 20	ohms ohms ohms ohm ohm ohm ohm ohm ohm ohm ohm ohm ohm	3.3 4 3.6 4 4.7 4 4.7 4 4.7 4 5.6 k 6.8 k 7.5 k 00. YRENE required 25 pf 30 pf 30 pf 39 pf ON SERS No AN'T GE AIN'T GE AIN'T GE AIL 1 5. 50 k 0.0 k SERS No SERS No	CONDENSE 50 pf 58 pf 62 pf 72 pf FRANY CHE 0.K	10 k ohm 16 k ohm 18 k ohm 22 k ohm 24 k ohm 24 k ohm 24 k ohm 30 k ohm ERS 10/- r 80 pf 82 pf 100 pf 125 pf EE TH Price EAPER! 1 1 ! ! 10/- per 100 10/- per 50 - each £2 dozen 0 LEAKS OR 0/200 1/- each. amp) £1 each.	39 43 47 51 62 75 82 135 pf 140 pf 158 pf 170 pf IESS GANC SKEL VOLU TELE 1 THIN CO-A: CRYS RECO ACOS ACOS TRAN	k ohm 99 k ohm 13 k ohm 36 k ohm 43 k ohm 47 k ohm 62 Total: Total: Iso pf 2 190 pf 3 200 pf 4 240 pf 6 E PRIC EE PRIC EE PRIC EE ON PRESETS ME CONTROLS VISION REMOT double pot., 5 resis CONNECTING KIAL CABLE. Bis TAL MIKES. 10/- RD PLAYER C/ GP67/2 15/- (Mon GP93/1 25/- (Com GP93/1 25/- (Com	1 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 3/- per doz 250 pf 250 pf 200 pf 20	1.5 1.8 1.8 3.6 5.1 6.2 7.5 en of any 580 pf 300 pf 320 pf 300 pf 320 pf 300 pf 22(6d. each - dozen. neg. with 0LS. Phin idensers, rds 1/-, 1 , £1 50 yd S P94/1 30/ COS GP9 ENT LIG 15 wa	meg ohm meg ohm meg ohm meg ohm meg ohm meg ohm one value. 1,000 pf 1,100 pf 1,500 pf 2,200 pf 2,200 pf 1,500 pf 2,200 pf 1,500 vds 1,500 vds 1,50	9.1 meg oh 10 meg ohr 2,500 pf 2,700 pf 3,000 pf 6,200 pf 6,200 pf 2,700 st 3,000 st 3,000 st 5,200 st 2/- each. 11' 7-way ca st £3/3/) 1,000 yds. 5 mic) ond needle 32 ond needle 37 DLT
ck the vi 13 o o 22 o 36 o 47 o 91 o 220 o 47 o 0 o 47 o 0 o 7 o 1 v 220 o 47 o 0 o 47 o 0 o 47 o 0 o 47 o 0 o 47 o 1 o 220 o 47 o 1 o 220 o 47 o 1 o 220 o 47 o 0 o 47 o 0 o 47 o 0 o 47 o 1 o 220 o 47 o 0 o 1 o 20 o 20 o 1 o 20 o 1 o 20 o 1 o 20 o 20 o 1 o 20 o 20 o 20 o 20 o 20 o 20 o 20 o 20	alues required. hms foo hms 750 of hms 1.5 k hms 1.5 k hms 1.8 k hms 2.2 k r selection (inixed) f MICA/CERAMIC n following values. 5 pf 12 6 pf 15 8 pf 18 10 pf 22 CD POLYESTER 3d. each 3d. each 3d. each 3d. each 6d. each 6d. each 6d. each 6d. each 6d. each 1/- each Unt lots of 100 per t unt lots of 100 per t UTPUT (Similar (PLANAR TRAN RTS. Gain of 20/5 s similar to OCP 71	ohms ohms ohms ohm ohm ohm ohm ohm ohm ohm ohm ohm ohm	3.3) 3.6) 4.7) 4.7) 4.7) 5.6 k 6.8 k 6.8 k 6.8 k 6.8 k 6.8 k 5.6 k 6.8 k 5.6 k 6.8 k 5.6 k 6.8 k 7.5 k 00. YRENE 7.5 k 1.0 k 1.	CONDENSE 50 pf 58 pf 62 pf 72 pf IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF IPAF	10 k ohm 16 k ohm 18 k ohm 22 k ohm 24 k ohm 24 k ohm 24 k ohm 30 k ohm ERS 10/- g 80 pf 82 pf 100 pf 125 pf EAPER! 1 1 ! 1 10/- ger 10/- ger 10/- ger 10/- ger 10/- ger 10/- ger 10/- ger 10/- ger 0 LEAKS OR 0/2 dozen, 24/- dozen, 24/- dozen, 53'' 900' 9/-	39 43 47 51 62 75 82 140 pf 158 pf 170 pf IESS GANC SKEL VOLU TELE 1 THIN CO-AL CRYS RECO ACOS ACOS ACOS TRAN 8 watt	k ohm 99 k ohm 13 k ohm 36 k ohm 43 k ohm 47 k ohm 62 Total: Total: Iso pf 2 190 pf 3 200 pf 4 240 pf 6 E PRIC EE PRIC EE PRIC EE ON PRESETS ME CONTROLS VISION REMOT double pot., 5 resis CONNECTING KIAL CABLE. Bis TAL MIKES. 10/- RD PLAYER C/ GP67/2 15/- (Mon GP93/1 25/- (Com GP93/1 25/- (Com	1 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 0 k ohm 3/- per doz 50 pf 20 pf 20 pf 20 pf 50 pf 50 c CCES 00 pf 50 c CCES 00 c 0 k ohm 50 pf 20 pf 50 c CCES 00 c 0 k ohm 50 c 0 k ohm 50 c CCES 00 c 0 k ohm 50 c 0 k ohm 50 c 0 k ohm 50 c 50	1.5 1.8 1.8 3.6 5.1 6.2 7.5 en of any 580 pf 300 pf 320 pf Total: 12 2/6d. each - dozen. heg. with OLS. Phi densers, rds 1/-, 1 5 st 50 yd COS GP9 COS COS COS GP9 COS COS COS COS COS COS COS COS COS COS	meg ohm meg ohm meg ohm meg ohm meg ohm meg ohm one value. 1,000 pf 1,100 pf 1,500 pf 2,200 pf 2,200 pf 2,200 pf 2,200 pf 1,500 yf 2,200 pf 1,500 yf 2,200 pf 3,100 yds 7/6d., is. • (Stereo, ccra 3,11 with diam 4/1 with diam HTS. 12 V(tt 18 [°] tube, B tage 3,- KIT	9.1 meg oh 10 meg ohr 2,500 pf 2,700 pf 3,000 pf 6,200 pf 6,200 pf 2,700 st 3,000 st 3,000 st 5,200 st 2/- each. 11' 7-way ca st £3/3/) 1,000 yds. 5 mic) ond needle 32 ond needle 37 DLT

Wireless World, January 1970



Amplifier Kits

Styled and kitted by T.R.S., using quality components, including valves or transistors and excellent instructions. Backed by T.R.S. service.

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MULLARD 2-VALVE PRE-AMP with switching for 5 inputs; bass/treble/volume controls, etc. Sensitivity at Input—4mV max. to 330 mV into 80K-1 Megohm; Response 20-25,000 Hz +1dB. KIT £6.19.6; BUILT £9.10.0 (Carr. either, 5/6).

MULLARD 10.10 STEREO AMPLIFIER. Input sensitivity—210 mV per ch.: Response 12 Hz—35KHz + 3d8: 10 watts R.M.S. output per channel into 3 or 15 ohms. KIT 418.10.0: BUILT 422.10.0 (Carr. either, 126.) As above, less controls and panel. Kit 417.0.0; Built 421.0.0 (2+2 pre-amp. essential).

2+2 STEREO PRE-AMP similar to Mullard 2-valve pre-amp, but doubled with gang controls and balance. BUILT £13.19.6 (Carr. 7/6).

controls and balance. BUILT 213.19.6 (corr. 7/o). T.R.S. 4.4 STEREO AMP Low cost transistor amplifier based on Mullard modules. 4.44 watts output. For B-low the strain state of the state assembly. Amo and pre-amp with front panel and knobs. Kit 27.19.6 (Corr. 3/6); Teak sided Cabinet 21.17.6 (Corr. 2/6); 24V. Power pack 22.5.0 (Corr. 2/6); Complete kit Inc. DIN plugs and sockets 612.10.0 (Corr. 7/6).

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P.W. 12-12

P.W. 12-12 T.R.S. have produced their own klt version of this outstandingly good combined stereo amp and pre-amp. It conforms closely to Prac-tical Wireless's excellent circuit but Is styled for a flatter, more conventional cabinet which will be shortly available. Klt Includes two-tone front panel and control knobs. Inputs-Mag. P.U. (R.I.A.A.) 2.5mV into 68 Kohms; Ceramio-Radio: Response-20Hz to 30KHz + IdB. Out-put-12 watts per ch. R.M.S. into 15 ohms. Power/Amp/Pre-omp Kits ovoilabl

Power/Amp/Pre-amp Kits available separately.

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 10 ditto
 1.2 Meg-10 Meg.

 10 ditto
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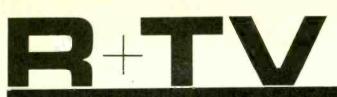
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SPECIFICATION:-R.M.S. power output: 3 watts per channel into 10 ohms speakers INPUT SENSITIVITY: Suitable for medium or high output crystal cartridges and tuners. Cross-talk better than 30dB at 1Kc/s. CONTROLS: 4-position selector switch (2 pos. mono and 2 pos. stereo) dual ganged volume control. TONE CONTROL: Treble lift and cut. Separate on off switch. A preset

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Sensitivies for 10 watt output at 1 KHz into 3 ohms. Tape Head: 3mV (at 3½ i.p.s.). Mag. P.U.: 2 mV. Cer, P.U.: 80 mV. Tuner: 100 mV. Aux. 100 mV. Tape/Rec. Output: Equalisation for each input is correct to within ±24B (R.I.A.A.) from 20 Hz to 20KHz. Tone Control Range: Bass ±13 dB at 60 Hz. Treble ±14 dB at 15 KHz. Total Distortion: (for 10 watt output) <1.5%. Signal Noïse: <-60dB. AC Mains 200-250v. Size 12½" long. 4½" deep. 2½" high.



The Viscount INTEGRATED HIGH FIDELITY TRANSISTOR STERED AMPLIFIER f14 5s. + 7/6 p. & p.

SPECIFICATION

OUTPUT: 10 watts per channel into 3 to 4 ohms speakers (20 watts) monoral. INPUT: 6-position rotary selector switch (3 pos. mono and 3 pos. stereol. P.U. Tuner. Tape and Tape Rec. out Sensitivities: All Inputs 100 mV into 1.8M ohm.

These 5 items can be purchased together for £29 10s+£1 10sp. & p.

PREQUENCY RESPONSE: A0Hz-200KHz±20B. TONE CONTROLS: Separate bass and treble controls. TREBLE 13dB lift and cut (at 15KHz) BASS: 15dB lift and 25dB cut (at 50Hz).

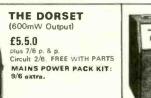
VOLUME CONTROLS: Separate for each channel. AC MAINS INPUT: 200-240v. 50-60Hz. Viscount Mark II for use with magnetic pick ups specification as above. Fully equalised for magnetic pick ups. Suitable for cartridges with minimum output of 4mV/cm/sec. at 1kc. Input magnetic pick ups. Suitable for cartrioges Impedance 47k. £15 15s. plus 7/6 p. & p.



SPECIFICATION

SPECIFICATION

OUTPUT: 10 watts into a 3 ohms speaker. INPUTS: (1) for mike (10 m.v.). Input (2) for gram. radio (250 m.v.) individual bass and treble control. TRANSISTORS: 4 silicone and three germanium.



7-transistor fully tunable M.W.-L.W. superhet portable with baby alarm facility. Set of parts. The latest modulized and pre-alignment techniques makes this simple to build. Sizes: 12" x 8" x 3





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General Purpose Amplifier In teak-finished case

MAINS INPUT: 220/250 volts. SIZE: 104'' x 44'' x 24''. MIKE TO SUIT (CRYSTAL): 12/6d. 1/6d. p. & p.

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8" x 5" speaker 14/61. 4 3/- p. & p

Solid State

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STEREO PRE-AMPLIFIER **SIGNED THE ANY PLIFIER** Inputs—6 position rotary switch (3 position mono, 3 position stereo). Tuner 150 mV time 800k. Magnetic pickup fully equalised and suitable for magnetic cartridges with minimising output of 4mV/mvise. Load 47k. Ceramic pickup 150 mV into 80kb. Sensitivities taken for 200mV output. Controls—separate volume controls for each channet. Twin ganged bass, 12d8 fit and 15d8 cut at 60c/s. Twin ganged treble. T0d8 lift and 15d8 cut at 10kc/s. Voltage required 23.30v. DC at SmA. Size 12²; a 3] * a 2]². In teak finished case. complete with front panel and knobs. Built and tested **f7.7.0** plus 5/- p. 8 p.

X101 10w. SOLID-STATE HI-FI AMP

With Integral Pre-amp. Specifications: Power Output linto 3 ohms speaker) 10 wetts: Sensitivity (for rared ourput): 1mV into 3K ohms (0,33 microam) Ortal Distortion (art IKH); At 5 wetts: 0.35%; At rared output 1.5%, Frequency, Response: Minus 3 dB points 20 Hz and 40 KHz Speaker: 3-4 ohms 13-15 boms may be used! Supply vottags: 24v D.C. at 80D mA, (6:24v may be used!

P101 M (mono) 35/- p. & p. 4/6: P101 (stereo) 42/6 p. & p. 4/6.





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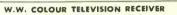


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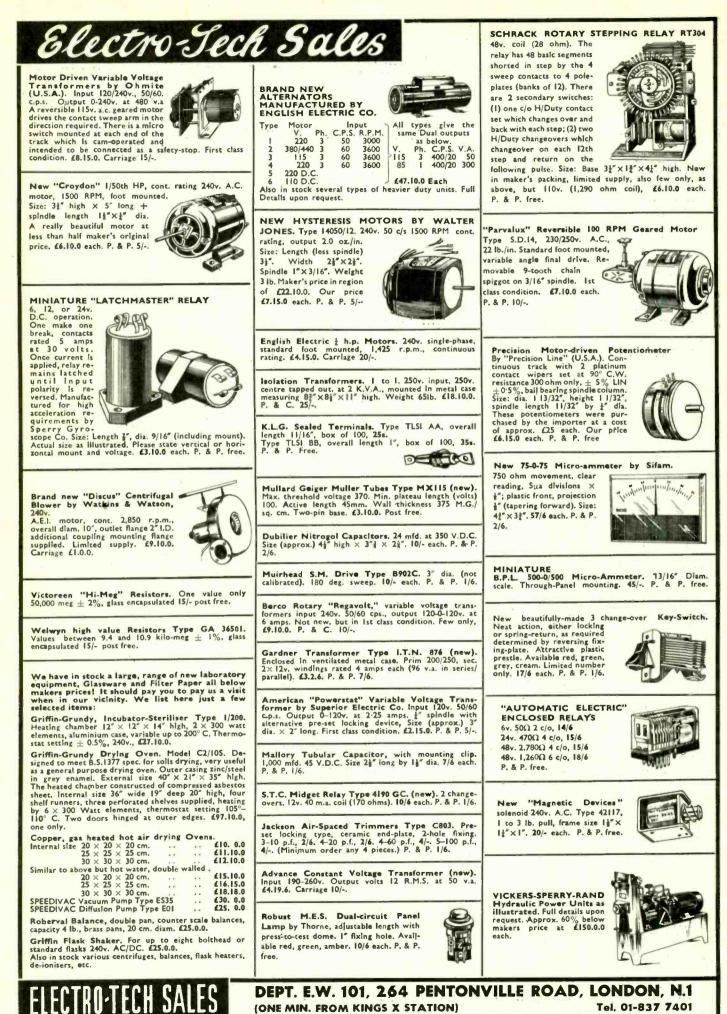
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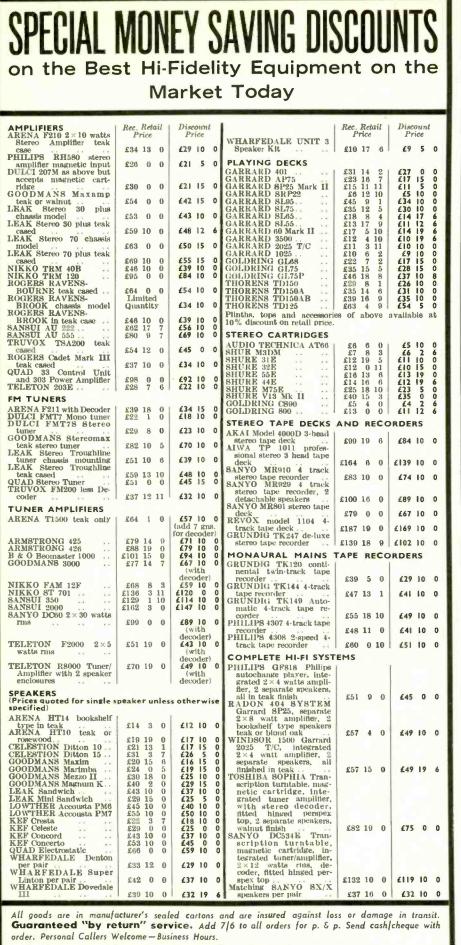
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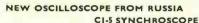
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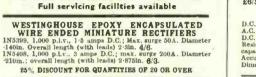


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2691

TECHNICIAN required for **ELECTRONICS WORKSHOP**

Applicants should have a minimum of two years' experience of construction and/or servicing of electronic equipment using transistors and integrated circuits, and some transistors and integrated circuits, and some knowledge of digital and pulse techniques. Qualifications to ONC or Intermediate City and Guilds techniclans level. Opportunities for day release. Salary according to age, qualifications and experience. Apply to the Administrative Secretary, Medical Research Council's Laboratory of Molecular Biology, Hills Road, Cambridge.

2731

NORTH-EAST ESSEX TECHNICAL COLLEGE

Sheepen Road, Colchester

DEPARTMENT OF ELECTRICAL ENGINEERING

LECTURER GRADE II

is required to teach TELECOMMUNICATION SUBJECTS up to first degree standard. Applicants must have a University degree in Electrical Engineering or Physics with suit-able teaching and/or industrial experience. Salary: Lecturer Grade II-£1,827-£2,417 p.a. Assistance with removal expenses may be considered.

Application forms are available from the Principal at the College, to whom they should be returned within fourteen days of the appearance of this advertisement. Please state ref. WW. 2730

Rank Strand Electric Limited, who are leaders in their field of theatre and television studio equipment, require two

electronics commissioning engineers

around £1,500 p.a.

To test and commission sophisticated lighting control systems into theatres and television studios at home and abroad.

Based with the Research and Development Unit at Brentford, Middlesex, they will initially complete a three-month training course before commencing operations in the field.

Applicants must have sound knowledge and experience of electronics. Ability to communicate effectively with customers Important. Knowledge of theatre and television an advantage

Must be willing to travel extensively and work irregular hours.



Wireless World, January 1970

APPOINTMENTS



a98

and work at the nerve centres of civil aviation

The National Air Traffic Control Service of the Board of Trade needs Radio Technicians to install and maintain the very latest electronic aids at Civil Airports, Air Traffic Control Centres, Radar Stations and specialist establishments. Vacancies exist in various parts of the United Kingdom.

This is responsible demanding work (for which you will get familiarisation training) involving communications, computers, radar and data extraction, automatic landing systems, and closed-circuit television. It offers excellent prospects with ample opportunities to study for higher qualifications in this fast-expanding field

If you are 19 or over, with at least one year's practical experience in telecommunications, fill in the coupon now. Preference will be given to those having ONC or qualifications in Telecommunications.

Salary : £985 (at 19) to £1,295 (at 25 or over); scale maximum £1,500 (higher rates at Heathrow). Some posts attract shift-duty payments. The annual leave allowance is good and there is a non-contributory pension scheme for established staff

Complete this coupon for full details and application form : To : A. J. Edwards, C. Eng., M.I.E.E., M.I.E.R.E., Room 705, The Adelphi, John Adam Street, London WC2, marking your envelope 'Recruitment'.

Name. Address ww/B3 Not applicable to residents outside the United Kingdom.

National Air Traffic Control Service

INTERTEL COLOUR TELEVISION

requires

ENGINEERS

in their Vision and Video Tape Departments to be based at their Dean Street, London, Studio. Applicants should have a good working knowledge of Colour Television Practice.

Applications to:

CHIEF ENGINEER INTERTEL COLOUR TELEVISION WYCOMBE ROAD, WEMBLEY MIDDLESEX

2726

NCR requires additional ELECTRONIC, ELECTRO **MECHANICAL ENGINEERS and TECHNICIANS to** maintain medium to large scale digital computing systems in London and provincial towns.

Training courses will be arranged for successful applicants, 21 years of age and over, who have a good technical background to ONC/HNC level, City and Guilds or radio/radar experience in the Forces.

Starting salary will be in the range of £900/£1,250 per annum, plus bonus. Shift allowances are payable, after training, where applicable. Opportunities also exist for Trainees, not less than 19 years of age, with a good standard of education, an aptitude towards and an interest in, mechanics, electronics and computers.

Excellent holiday, pension and sick pay arrangements. Please write for Application Form to **Assistant Personnel Officer** NCR, 1,000 North Circular Road, London, NW2 quoting publication and month of issue.

Plan your future with

Construction of Construction o

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APPOINTMENTS



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for the Zambia Broadcasting Services, Ministry of Information, Broadcasting and Tourism, on contract for one tour of 36 months in the first instance. Commencing salary Kwacha 3,408 (£Stg. 1988) rising to Kwacha 3,516 (£Stg. 2051) a year, plus an Inducement allowance of Kwacha 1002–1034 (£Stg. 585–£Stg. 603). A Direct Payment of £291 is also payable direct to the officer's bank in the U.K. Salaries are subject to upward revision with effect from 1st January 1970. Gratuity 25% of total salary drawn. Both Gratuity and Direct Payment are normally TAX FREE. Free passages. Accommodation at moderate rental. Education allowances. Liberal leave on full salary or terminal payment in lieu. Contributory pension scheme available in certain circumstances.

Candidates, between 25-55, must have passed City and

Guilds final certificate in Telecommunications or equivalent and should have had at least eight years experience with a broadcasting organisation, with particular experience in the installation of recording equipment and studio control equipment.

The officer will be required to maintain and service audiovisual aid equipment and install and operate public address/ recording film projection equipment when and where required. He will be required to supervise workshops and staff in the absence of the Senior Maintenance Engineer.

Apply to CROWN AGENTS, 'M, Division, 4 Millbank, London, S.W.I., for application form and further particulars stating name, age, brief details of qualifications and experience and quoting reference number M2Z/691029/WF.

2692

EXETER AND MID-DEVON HOSPITALS Management committee group works department

Technician–Medical Physics Grade II,

for maintenance of dialysis machines and electronic equipment. Salary £1,313 × (8) to £1,671. Qualifications required—H.N.C. or H.N.D. in Electrical Engineering or equivalent.

Applications together with names of three referees to reach the Group Engineer, 26 Queen Street, Exeter, Devon, within 10 days. 2769

COMPUTER STAFF

Vacancies exist in our Computer Services Dept. for EQUIPMENT CONTROLLERS to be responsible for the installation and performance monitoring of three IBM360/50 Computers used for real-time data processing. Applicants should have some knowledge of digital computers and may at present be employed as commissioning engineers or in a similar post. Knowledge of teleprocessing, Datel services and computer terminal equipment would be an advantage though it is not essential as full training will be given.

Salary up to £2000 according to experience and ability. Assistance with house purchase is available.

Please apply in writing giving full particulars to : P. Jordan. Computer Services Manager, CENTRE-FILE LTD., Park House, 16 Finsbury Circus, London, E.C.2. 2763 **GEC-Marconi Electronics**

Technicians and Engineers for St. Albans and Luton *qualified or not* !

Vacancies in all grades

- VACANCIES exist for work on testing and calibrating valve and solid-state electronic measuring equipments embracing all frequencies up to u.h.f. in Production, Service and Calibration departments.
- **APPLICATIONS** are invited from people of all ages with experience or formal training in electronics and from ex-Armed Services technicians.
- SALARIES up to £1,600 negotiable and backed by valuable fringe benefits.
- **RE-LOCATION** EXPENSES available in many instances.
- CONDITIONS excellent: free life assurance, pension schemes, canteen, social club.
- 37¹/₂-hour, 5-day, office-hours week.
- WRITE or 'phone Personnel Department stating age, details of previous employment, training, qualifications, approximate salary required.





Longacres, St. Albans, Herts. Tel: St. Albans 59292 Luton Airport, Luton, Beds. Tel: Luton 31441 A GEC-Marconi Electronics Company

APPOINTMENTS

Product Test Technicians

Career Opportunities with IBM Manufacturing

We need high calibre men to fill vacancies created by promotion and programme expansion.

The job

Is to commission the latest IBM products and systems in production at the Scottish plant, near Greenock, and requires an intimate knowledge of the equipment under test, which can include computers, punched card and tape peripherals, magnetic disk and tape storage, high and low speed printers, visual display units, multiplexors, Teleprocessing and optical character recognition equipment. The products have to be tested thoroughly, and all faults traced and rectified. The work is interesting and absorbing, and the prospects for the right man are good.

Training

Will be a mixture of formal and "on the job" instruction. We will teach you all you need to know about IBM equipment – providing your basic knowledge is to the required level.

Pay and conditions

Starting salaries will be excellent.

Benefits include a non-contributory pension, immediate free life assurance and full sickness pay for up to 26 weeks in any 12 months. The 254,000 square feet plant is modern and situated in a pleasant rural valley. There is a subsidised restaurant.

Working conditions are excellent and there are good recreational facilities in the area. IBM will assist with removal expenses where applicable.

The man

Will be at least 18 and probably less than 30 and have a strong electronic background, with experience in, for example, the testing of electronic products, maintenance of radio, radar or TV or similar work in the armed forces.

He will probably have, or be near to attaining, a qualification such as HNC, ONC, first class PMG, final RTEB, or final City and Guilds (Course Nos. 47, 48, 49, 57, 300). A knowledge of transistor circuitry and the use of oscilloscopes will be a distinct advantage.

If you have what we need, and are keen to join a vigorous, expanding and up-to-the-minute industry, please write, giving details of your age, experience and qualifications, and quoting ref. No. PT2/WW/169 to: Manager, Personnel Selection, IBM United Kingdom Limited, P.O. Box 30, Spango Valley, Greenock.



for

TRAWSFYNYDD NUCLEAR POWER STATION Trawsfynydd, Merionethshire, North Wales.

The work involves the diagnosing of system faults, carrying out special investigations, originating, modifications and designing special test equipment. The successful applicant will also be expected to assist in the training of Maintenance Craftsmen. Some experience within this field would be an advantage.

Applicants must possess academic qualifications leading to Graduate Membership of the I.E.E. or I.E.R.E.

Salary is within the range £1,500 – £1,880 per annum. Superannuable.

Conditions of service are in accordance with the National Joint Board Agreement for the Electricity Supply Industry.

Applications in writing stating age. experience etc.. to Regional Personnel Manager, 825 Wilmslow Road, East Didsbury, Manchester M20 8RU to arrive not later than 30th December 1969. Please quote Vacancy No. E.566/567/W.

2696

University of London Audio-Visual Centre

TELEVISION ENGINEERS and JUNIOR TELEVISION ENGINEERS

to work on maintenance and operational duties with its studio and mobile equipment, under the supervision of the Chief Engineer, R. H. Bradley, MBE.

Applicants for the post of Television Engineer should have technical experience in broadcast or educational television. A broad knowledge of other audio-visual equipment would be an advantage.

Applicants for the Junior posts should be 18-20 years old, with a basic knowledge of electronic equipment and some experience in its use. Formal technical or educational qualifications would be an advantage.

Salaries :

TV Engineer. Starting at £1600-£2000 according to qualifications and experience. University Pension Scheme. Junior TV Engineer. Starting salary £800-£1000 according to qualifications.

Applications to the Director: University of London Audio-Visual Centre 11 Bedford Square, London, WC1

www.americanradiohistory.com

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APPOINTMENTS

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11) Europ Europ

Meteorological Department

requires

Sectional Engineer Grade II (Telecomms.)

To serve on contract for one tour of 21-27 months in the first instance. Salary in scale EA.Shg. 24300-27780 (approx. $\pounds S.1417-1620$ p.a.) plus an Inducement Allowance normally tax free, of $\pounds S.822-886$ p.a. paid direct into officer's bank in U.K. Gratuity 25% of total emoluments. Generous paid leave. Education Allowances. Furnished accommodation at reasonable rental. Free passages. Contributory pension scheme available in certain circumstances.

Candidates, up to age 45, must possess O.N.C. or City and Guilds Final Certificate (Telecomms) plus 7 years relevant experience in telecomms. engineering. Equivalent experience in one of the armed services is acceptable. Candidates must have a good theoretical and practical knowledge of FSK, ISB and SSB receivers and transmitters and of Mufax and facsimile transmitters and recorders. A good working knowledge of radar systems is essential.

The officer will be responsible to the Chief Sectional Engineer for the operation and maintenance of the Department's radio telecommunications, radio-sounding and radar equipment. He will be liable for service anywhere in East Africa but will probably be stationed at Entebbe, Dar es Salaam or Nairobi.

Apply to CROWN AGENTS, 'M' Division, 4 Millbank, London, S.W.I., for application form and further particulars stating name, age, brief details of qualifications and experience and quoting reference number M2K/690413/WF

TECHNICIAN

required for calibration and servicing of flow meter instruments. Electronic experience required in television servicing or pulse techniques. Transistor experi-ence essential. Digital experience an advantage. Good commencing salary and prospects. Permanent pensionable position.

Apply: B. RHODES & SON LTD., Danes Road, Off Crow Lane, Romford, Esssex. Telephone Romford 62333/4/5. 2774

ELECTRONICS TECHNICIANS

SENIOR TECHNICIAN prototype wireman required for research contract. TECHNICIAN for interesting project and development work in research and teaching laboratories. Day release available.

Incremental salary ranges £868-£1252, £1151-£1486 depending on age, experience and qualifications. 371 hour week, good working conditions and holidays. Apply in writing to Mr. E. Thompson, Royal School of Mines, Mining & Mineral Technology Department, Prince Consort Road, London SW7 2764

Electronics Engineer

> Professional Audio, Video and Instrumentation Equipment.

3M Company, one of the world's foremost names in magnetic recording, is ex-panding distribution of its "Mincom" range of equip-ment in the audio, video and instrumentation fields in the United Kingdom.

An interesting opportunity is available for a young man aged 23-26 to join us as an Electronics Engineer. He must possess a good electronics background, possibly obtained in the radio com-munications field. Ideally he should be qualified to C & G or ONC/HNC standard but lesser qualifications may be acceptable in the case of an

applicant with exceptional practical experience.

The position will be London based but ultimately the successful applicant will be re-quired to undertake some travelling working on his own initiative on field assignments.

This is a progressive position in a fast growing field. The start-ing salary will be attractive and there are first class employee benefits. Please write in confidence with concise details, age, qualifications and experience to:

Mr. D. J. Stuckey (Ref. MT) 3M Company Ltd., 3M House, Wigmore Street, London WIA IET.

a102

TECHNICAL AUTHORS

We are one of the World's leading manufacturers of Flight Simulators which incorporate both analogue and digital computers. We require Technical Authors with a sound knowledge of electronics who preferably have some knowledge of basic digital computer operation. Authors will produce operating and maintenance manuals, and must be able to write literature in a clear and concise style. Formal qualifications are desirable, but not essential. There will be ample opportunity to employ a measure of creative expression. Simulation is based on novel applications of known techniques. This work is certainly not of a monotonous nature.

INSTRUCTORS

are required to lecture to customers and engineers on digital and analogue computing techniques at basic, intermediate and advanced levels. Students are required to maintain and programme highly sophisticated flight simulators. Applicants should preferably be qualified to at least H.N.C. level or equivalent. Preference will be given to those with practical analogue or digital computing experience, but applications from men with several years' industrial experience and/or who have a flair for and a genuine interest in lecturing will certainly be welcome. Training in the Company's advanced computing techniques will be given.

This Company offers good working conditions, welfare benefits. There is a contributory pension scheme coupled with free life assurance.

Apply to: H. C. Hall, Personnel Manager, REDIFON LIMITED Flight Simulator Division, Gatwick Road, Crawley, Sussex Telephone: Crawley 28811



ST. JOHN'S COLLEGE OF EDUCATION • YORK

Dept. of Closed Circuit Television

Applications are invited for the post of SENIOR TECHNICIAN, to join a team making television programmes for this and five associated colleges.

Duties will include the maintenance of cameras and videotape recording equipment. Opportunities for operational and production work will occur. A mobile recording van is in regular use and ability to drive would be an advantage.

Salary : Local Government Scales : Technical Grade 6 (at present £1,540-£1,775): the post is superannuated and good holidays are given.

Applications (no special form), should be made in writing to the Principal, stating qualifications, experience and the names of two referees. Closing date 31st December, 1969. 2762

SOUTH AFRICA

FULLY QUALIFIED **RADIO & TELEVISION TECHNICIAN**

Applicants should be capable of supervising a workshop from which the installation and repairs to all types of radios and television are undertaken.

Applications with full details of experience etc., should be sent in the first instance to Mr. E. B. UNWIN.

NEL & UNWIN (PTY) LTD. P.O. BOX 199, KROONSTAD SOUTH AFRICA

a103

APPOINTMENTS

CONTINUOUS

Standard Telephones & Cables, Microwave and Line Division based at Basildon are growing fast. In order to keep pace with this consistent growth rate we require the following [

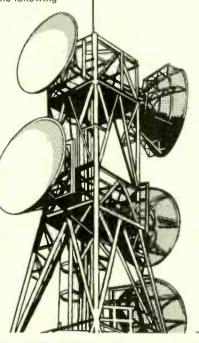
Installation Engineers Technicians & Testers

Ref. 25720

To test and commission Multiplex, Co-axial Line and Microwave Radio Systems.

Ideal candidates will be less than 45 years of age with practical experience on some of the above equipment. These challenging posts call for drive, initiative and common sense. It is necessary for applicants to be prepared to work anywhere in the U.K.

> Applications should be addressed to The Personnel Officer, STC Chester Hall Lane, Basildon, Essex.



Test Technicians Ref. 27221

The diversity of products manufactured at the Basildon Plant demands experienced testing staff for work on complex transmission systems.

Candidates should hold an ONC in electrical engineering and be able to offer considerable practical experience in the Tield of testing and fault clearing all types of land-unit, pcm and microwave equipment.



TECHNICIAN

required in APPLIED ACOUSTICS RESEARCH LABORATORIES situated near Fulham Broadway, S.W.6. Varied work, but a knowledge of electronic construction and design an advantage. Day release facilities for further study.

Salary: £868—£1,252 p.a., depending upon age, experience and qualifications.

Application forms and further information from the Superintendent of Laboratories (T.A.), Department of Physics and Electronics, Chelsea College, Manresa Road, London, S.W.3. 2735

EAST SUFFOLK COUNTY COUNCIL

Lowestoft College of Further Education Principal A. E. Boddy, B.Sc. (Econ.), F.R.G.S.

LECTURER GRADE 1

required for City and Guilds Radio and Television Servicing Mechanics' and Technicians' Courses, including colour television.

Applicants should have appropriate technical qualifications, together with suitable industrial and Teaching experience. Ability to offer teaching in similar courses would be an advantage.

The appointment is vacant as from the 1st April, 1970, but an earlier commencing date may be negotiated.

Salary in accordance with the Burnham Scale for Lecturers Grade J, £1,110 to £1,955, plus increments for approved qualifications and training. Starting point within the scale determined by previous Industrial and Teaching experience.

Applications should be sent as soon as possible to The Principal of the College, on application forms available from the Secretary, Lowestoft College of Further Education, St. Peter's Street, Lowestoft, Suffolk.

2768

SENIOR LABORATORY TECHNICIAN

A SENIOR ASSISTANT with a good understanding of electronics is needed to join a small team providing physics support to the Isotope Production Unit at Harwell. The team is mainly concerned with making accurate measurements of a wide variety of radiation sources and with the development and maintenance of the necessary measurement system. The post is tenable at Harwell.

QUALIFICATIONS & EXPERIENCE:-

The minimum age for appointment is 27 and the minimum qualifications necessary are four 'O' levels including English Language and Mathematics or a Science subject. Electronics experience is essential and experience in the measurement of radiation sources would be advantageous.

SALARY: £1,350 rising to £1,755

APPLY TO: The Personnel Officer

THE RADIOCHEMICAL CENTRE Amersham Bucks

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a104

Help keep aircraft on the straight and narrow

Air traffic has become so congested that complex electronic techniques are used as an aid in controlling aircraft both on airways and on airport approaches. As a Telecommunications Technical Officer III in the National Air Traffic Control Service of the Board of Trade, your job would be to install and maintain various air navigational and landing aids at civil airports, and communications and computer systems at radar stations and signals centres.

Because you handle such advanced equipment, you will receive thorough training. Study for higher qualifications is encouraged, and this could range from short courses with financial assistance to full-time study at a university or technical college.

Pay: (London rates—a little less elsewhere) £1,350 starting salary at 23, £1,625 at age 28 or over on entry, rising to £1,810. Within 3 years you could be upgraded, and on a scale rising to £2,050. A few years after that, you could be in the salary bracket going up to £2,375, and there are several higher grades still.

Qualifications: O.N.C. in Engineering, including a Pass in Electrical Engineering; or equivalent standard of technical education.

Send for full details and an application form (which must be returned completed by January 2nd, 1970) to: Civil Service Commission, 23 Savile Row, London, W1X 2AA. Please guote S/207/-.



SENIOR DEVELOPMENT ENGINEER.

WW.21. £1900.

Our clients seek electronics engineer of at least H.N.C. standard with additional qualifications in control or digital techniques preferably around 30 years and living, or prepared to work, in the Hertfordshire area. The applicants should have shown some administrative ability in project coordination, possibly in avionics, military control systems, or high speed digital control circuits. Please apply in writing and in strict confidence,



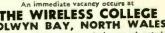
HENRY'S RADIO LTD. 303 EDGWARE ROAD, LONDON, W.2 HAVE THE FOLLOWING VACANCIES IN THEIR ORGANISATION

SALES ASSISTANTS

Young man with good general knowledge of electronic components required for our retail sales dept. Please telephone 723-1008/9 ext. 1.

SALES ASSISTANTS

Young man with a good general knowledge of HIGH FIDELITY EQUIPMENT required for our retail HI-FI SALES DEPT. Please contact MR. STEVENS, Telephone 723-6963. 2585



An immediate vacancy occurs at **THE WIRELESS COLLEGE COLUMYN BAY, NORTH WALES** for an additional instructor to assisting preparing students for P.M.G./M.P.T. examinations. The primary responsi-bilty will be the practical instruction on modern marines radio equipment. Applicants must hold a P.M.G. Certi-ficate and should have a sound technical knowledge. Recent marine operating and/or teaching experience is desirable but not essential. Write in the first instance to the Principal. 2776

SPECIAL OPPORTUNITY

Small company in medical electronics development (Richmond, Surrey) offers starting salary £1 300 to keen man suitable early directorship. Founder approaching retire-ment. Self-contained flat available. Phone 01-940 0865 evenings.

2766

SITUATIONS VACANT

A FULL-TIME technical experienced salesman re-quired for retail sales; write giving details of age, previous experience, salary required to -The Manager, Henry's Radio, Ltd., 303 Edgware Rd., London, W.2.

quired for retail sales; write giving details of age, Henry's Radio. Ltd., 303 Edgware Rd., London, W.2. [67]
 Are YOU INTERESTED IN HI FI? If so, and You Trade, an excellent opportunity awaits you at Telesonic Ltd., 243 Euston Road, London, N. W.1. Tel. 01-3877467. [21]
 Chintall and manage broadcasting equipment at new Radio School to be established at Choluteca, Republic of Honduras. Challenging post for the adventurous. Frying to get out of the rul! Volunteer terms: fares. board. logging, pocket-money, allowances. — Wite ; CHIE/OV, 38 King Street, London, W.C.2. [2737]
 Chief Electronics Workshops developing and anintaining electronic equipment for use in the teach-ing and research laboratories of the Departments of Electronics and Physics. Salary £1,510—£1,703 p.a. according to age and experience. Further information and application form from the Laboratory Superin-tender, Departments of Physics and Electronics. Chelsea College. Manresa Road, London, S.W.3. [2775]
 Chours TELEVISION. Multi-National Advertising Pageory require a Technical Assistant, or Assistant engineer, to operate and maintain colour television film scanning equipment. Candidates should be between the ages of 22 and 30 and should have "C" and "G" Telecomms. or equivalent, and preferably with experience on this type of equipment. Salary is noto. Ltd., 40 Berkeley Square, WIX 6AD. [2733]
 Check Dondrick TECHNICIAN required to assist in the for ontis instruments used in blochemistry. Applicants should possess O.N.C. or equivalent and be able to construct and test equipment from circuit diagrams. Salary according to sage and experience in the tange of solary commensure with and experiment of Bio-college. London, S.W.7. [2778]
 Chectronnic TECHNICIAN required to assist in the for octain gualifications. Superannualion scheme. Construct and test equipment from circuit diagrams should possess O.N.C. or equivalent and be able to construct and test equipment fro

Institut 242. [2760] I.-942 242. [2760] I.-B. Sharkar, and development. ELECTRONIC I.-ENGINEERS AND TECHNICIANS. Advanced optical/electronic systems to process text and pictures for publication are being developed in the Information Science and Technology Unit's laboratory. There are several immediate opportunities for: QUALIFIED ELECTRONIC ENGINEERS to join a team working on the design and development of systems using computer, CRT, TV, and optical techniques. Expert knowledge in at least one of these is essential. Salaries within the Development Officer scale, ranging from £1.500 to £3,500. LABORATORY TECHNICIANS for layout, construction and testing of prototype electronic apparatus. Salary scale up to £1.675. Write for application form to the Director of R & D. IPC Laboratories, Wood Lane, Hemel Hempstead, Herts. [2748]

QUALIFIED ENGINEERS considering emigration to Australia 1970 required to join new and vigorous firm of communications consultants. Interviews London and Enruingham. Brief resume to Box W.W. 2627. Wireless

APPOINTMENTS

ww/d22/jan

ELECTRONICS TECHNICIANS

this is the business machine explosion

Turnover doubled since 1962—now growing faster In the decimalization rush. Burroughs dominate the U.K. market in the new termInal computers and accounting machines. A wide variety of machines, an expanding market and a policy of promotion from within—all mean exciting opportunities, for trained electronics engineers, to develop their skills into the computer field or into the supervisory grades and beyond. Join the Burroughs boom—and grow with us.

If you are between 20 and 30, with an electronics qualification and want to train as a computer engineer—then there's a job for you with us. With Burroughs, you can find the free-

> Visible Electronic Accounting Calculators Cash Record Accounting & Adding

Burroughs

dom to develop your talents, open fresh horizons, learn new skills—on the largest third-generation systems in the World —these are the exciting prospects at Burroughs. In return, we're offering you 3 weeks' paid holiday, free life assurance and a contributory pension scheme.

a105

Take a big step now into one of today's development industries—fill in the coupon and send off for one of our application forms. The address is:

Bob Timms, Personnel Officer, Burroughs Machines Limited (Z), Heathrow House, Cranford, Hounslow, Middlesex.

NAME.....

ADDRESS.....

Hampstead High Fidelity

require the services of a qualified

SERVICE ENGINEER He must be conversant with the repair and service of quality High Fidelity equipment, a useful asset would be experience with CCTV and Colour TV. The man we are seeking must be very conscientious, adaptable and prepared to undertake occasional field work. He will be expected to set up and organise a completely new service department and control a small staff.

Applications are Invited in writing giving full details of qualifications, experience, etc., and marked for the attention of P. A. Rispoli, Esq., Hampstead High Fidelity, 91 Heath Street, London, N.W.3. 2746

UNIVERSITY OF ESSEX DEPARTMENT OF PHYSICS SENIOR TECHNICIAN

required for maintenance of electronic equipment, supervision of equipment in a teaching laboratory and assistance to research groups. Candidates should preferably have H.N.C. or equivalent qualification in electronics and experience with modern electronic circuitry and equipment. Salary range £1,056-£1,311 with additional allowance for approved higher qualifications.

Applications to the Registrar, University of Essex, Wivenhoe Park, Colchester, Essex.

2732

V.H.F. TELEVISION RELAY & COMMUNAL AERIAL SYSTEMS

We are planning a considerable expansion of our activities and have the following vacancies:

I. A SENIOR ENGINEER

to have control of all aspects of systems design, planning, estimating, installation and commissioning.

II. ENGINEERS

capable of undertaking either:

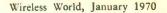
- (a) System planning and estimating.
- (b) control of installation work.
- or (c) test and commissioning duties.

Candidates for these appointments must have a good background of practical experience in this field of work, and an up-to-date knowledge of techniques and equipment.

Applications, which will be treated in strict confidence, should be sent to:

BRITISH RELAY

The General Manager, Special Services Division, British Relay House, 41, Streatham High Road, S.W.16



REDIFFUSION Television Service Management

Applications are invited for vacancies in the North East Region from mature T.V. Service Managers who are experienced in the operation of a large scale rental service organisation.

Applicants must be technically competent and have a strong flair for the solution of organisational, administrative and personnel problems.

An attractive salary will be offered and a car provided. Applications please to:

> The Chief Engineer, Rediffusion (North East) Ltd., Rediffusion House. Forth Banks, Newcastle upon Tyne, NE1 3RX.

> > 2683

BRISTOL POLYTECHNIC

FACULTY OF APPLIED SCIENCE Applications invited for the following post, duties to commence as soon as possible-

SENIOR TECHNICIAN IN PHYSICS

(Grade T.3)—Ref. No. T698/66/2 to take charge of Nuclear Physics/Radiochemistry Laboratory.

Applicants should be over 21 and have qualifications to at least O.N.C. or C. & G. Ordinary Technician standard and previous laboratory experience. 38-hour, 5-day week with generous holiday and sick pay schemes. Opportunities for evening work with additional pay. Permanent posts with superannuation under Local Govern-ment conditions of service.

Salary Scale: £930-£1,095. Starting salary dependent upon age, qualifications and experience. An additional £50 or £30 per annum will be paid for appropriate National Certificate and C. & G. qualifications.

Further particulars and application forms (to be returned within fourteen days of this advertisement) from Chief Administrative Officer, Bristol Polytechnic, Ashley Down, Bristol BS7 9BU.

Please quote Ref. No. T698/66/2 in all communications.

2767



Customer Services Department Engineers ervice

These appointments will be of interest to electronic engineers who have attained High National Certificate standard.

The Department is responsible for the installation and maintenance of high-power electronic and mechanical equipment, mainly on customer premises, both in the U.K. and abroad.

These monthly staff appointments offer an attractive salary and, in addition, the full use of a Company car.

The Company operates a generous contributory pension scheme, with a holiday entitlement of three weeks per year.

Applications should be made by letter or telephone to the Customer Services Manager.

LTV Ling Altec Limited,

Heath Works, Baldock Road, Royston, Herts. Tel: Royston 2424

2699

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RADIO SCHOOL IN PANAMA: Radio Technician required to operate radio school in Santiago de Veraguas. The school provides an elementary adult education programme and is now unable to function for lack of a technician to take charge. Challenging opportunity to fill a vital need in the development of rural areas. Volunteer terms: board, lodging, pocket-money, fares, allowances. Write CIIR/OV 38 King Street, London, W.C.2.

REDIFON LTD. require fully experienced TELE-COMMUNICATIONS TEST ENGINEERS and ELECTRONICS INSPECTORS. Good commencing salaries. We would particularly welcome enquiries from ex-Service personnel or personnel about to leave the Services. Please write giving full details to-The Personnel Manager, Redifon Ltd., Broomhill Road, Wandsworth, S.W.18. [26]

Wandsworth, S.W.18. [26 SENIOR ELECTRONICS TECHNICIANS are required by the Department of Applied Psychology to support the development of instrumentation for research work. Opportunities available for enterprising Technicians to gain experience in any of the following fields: Electro-physiological recording, magnetic tape systems, analogue computing, digital circuiting and building special peri-pherals for the departmental on-line PDP 9 computer. Experience and interest in building special peri-pherals for the department of the above fields essen-tial. Salary on the scale £1.026 to £1.280 per annum. With supplementary payment up to £80 per annum if approved qualifications held. Five day week. Pension Scheme. Application Forms from the Staff Officer, The University of Aston in Birmingham, Gosta Green, Birmingham 4, quoting reference L/515/W.W. [2739

UNIVERSITY OF SHEFFIELD: CHIEF TECHNICIAN required in Department of Chemistry to take charge of Electronics Workshop, concerned with development and construction of new electronic equipment for re-search and teaching, and maintenance and repair of wide range of electronic equipment. Experience neces-sary, qualifications preferable. Salary £1,385-£1,578 p.a. Write, stating age, qualifications and experience, to the Bursar (Ref. B.390), The University, Sheffield S10 2TN. [2741]

WE HAVE VACANCIES for Four Experienced Test Engineers in our Production Test Department. Applicants are preferred who have Experience of Fault Finding and Testing of Mobile VHF and UHF Mobile Equipment. Excellent Opportunities for promotion due to Expansion Programme. Please apply to Personnel Manager, Pye Telecommunications Ltd., Cambridge Works. Haig Road, Cambridge. Tel. Cambridge 51351, Extn. 327.

ARTICLES FOR SALE

A R88 mains transformers, £2.10.0; 12/24V carbon pile voltage regulator units. 30/-; 40 amp. 4-way terminal blocks. 5/-; Rotax rotary inverters, input 24V dc. output 115V. 3 phase, 400 Hz. 18 amps., £7.10.0; 200 amp. terminals, insulated heads, 8/6. Carriage and packing extra. Also, high voltage capacitors and in-sulators in stock. Westover Electronic Man. Co., Braidley House, St. Paul's Lane. Bournemouth. Tel. 23944. [2771

23944. **B**RAND NEW ELECTROLYTICS, 15/16 volt, 0.5, 1, 2, 5, 8, 10, 20, 30, 40, 50, 100, 200 mfds. 8d. Carbon Film Resistors 4 watt 5% E12 Series 10 ohms to 1 Megohm 1/6 dozen, minimum order 7/6, postage 1/-, The C. R. Supply Co., 127 Chesterfield Rd., Shefield S.8, [2747]

BUILD IT in a DEWBOX quality plastics cabinet. Diagood Rd., FERNDOWN, Dorset. S.A.E. for leaflet. Write now-Right now.

CLOSED CIRCUIT TV Mobile Control Room for sale. Purchased January 1969 and built by Ampex. The unit includes 3 Vidicon Cameras by Marconi and full sound and communication facilities. There is also a helican scan V.T.R. unit and an audio tape recorder on board, both built by Ampex. Offers should be made to Television Facilities, Queens Hall, Leeds 1. Loan Finance available.

COPIES of "Wireless World" from 1954 to 1968; offers please.—Grant, The Bungalow, 93 London Road, Hurst Green, Sussex. [2743]

DISC RECORDING UNIT complete with Leak stereo amplifier, mod. meters, console, motorised swarf suction unit. spare blank, sapphires, etc., etc. Sugden cutter unit. Good condition. Best offer over £50.—John King (Films) Limited, Film House, 71 East Street, Brighton, BN1 1NZ. [2749

FERRANTI. AF3. AF5, B1, OP2, OPM1(2), OPM5, OPM6, OPM1C, X17, £5 o.n.o. CARRIAGE PAID.-Norwood, May Hill, Ramsey, I.o.M.

FOR SALE BY TENDER 170,000 Valves

The Commissioners of CUSTOMS AND EXCISE are offering for sale by competitive tender approx. 170,000 valves in lots of approx. 1,000.

For further particulars, apply in writing to:

The Officer, Customs and Excise, Queen's Warehouse, Custom House, Lower Thames St., London, E.C.3, before January 2nd, 1970. 2751

HOW to Use Ex-Govt. Lenses and prisms. Booklets. Nos. 1 & 2, at 2/6 ea. List Free for S.A.E. H. W. ENGLISH, 469 RAYLEIGH RD., HUTTON, BRENT-WOOD, ESSEX.

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Ampex Quality Control Department now has vacancies for technicians to be responsible for fault finding and testing a wide range of Professional Audio and C.C.T.V. Magnetic Recording Equipment.

Experience gained in the electronic industry, radio or television servicing would be an advantage or a qualification of O.N.C. standard.

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Please write or telephone the **Personnel Officer**, **Ampex Electronics Limited**, Acre Road, Reading, **Berkshire**. **Telephone Reading 84411**.



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PRODUCTION TEST ENGINEER

With experience of valve and transistor audio equipments, wanted to join our senior staff. 40 hour week. Salary £1,200 per annum. ASSOCIATED ELECTRONIC ENGINEERS LTD., Dalston Gardens, Stanmore, Middlesex. Tel.: 01-204 2125 2729

Bath University of Technology



is required in the School of Mathematics to assist mainly in servicing and developing ANALOGUE AND DIGITAL COMPUTING devices.

Candidates should have experience in electronics, should possess a basic qualification and be competent in elementary workshop skills.

Salary in the range of £773— £1,077 per annum, according to age, experience and qualifications. Further details and application form from Registrar (S), The University, Bath, BA2 7AY, quoting ref. 69/83.

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for small shop producing transformers, 20vA-10KvA. Must be fully experienced, all stages of production. Commencing salary about £1,200.

Also Young Assistant with testing experience

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This private College provides efficient theoretical and practical training in the above subjects. One-year day courses are available for beginners and shortened courses for men who have had previous training. Write for details to: The Secretary, London Electronics College, 20 Penywern Road, Earls Court, London, S.W.5. Tel.: 01-373 8721. 84

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There will be a number of vacancies in the Composite Signals Organisation for experienced Radio Operators in 1970 and in subsequent years.

Specialist training courses lasting approximately nine months, according to the trainee's progress, are held at intervals. Applications are now invited for the course starting in September 1970.

During training a salary will be paid on the following scale:

Age 21		£800 per	annum
,, 22		£855	
,, 23		£890	
,, 24		£925	
,, 25	and over	£965	••

Free accommodation will be provided at the Training School.

After successful completion of the course, operators will be paid on the Grade 1 scale :

Age	21		£965	per annum
	22		£1025	,,
	23		£1085	
,,	24		£1145	
,,	25	(highest age point)	£1215	

then by six annual increases to a maximum of £1,650 per annum.

Excellent conditions and good prospects of promotion. Opportunities for service abroad.

Applicants must normally be under 30 years of age at start of training course and must have at least two years' operating experience. Preference given to those who also have GCE or PMG qualifications.

Interviews will be arranged throughout 1970.

Application forms and further particulars from :

Recruitment Officer, Government Communications Headquarters, Oakley, Priors Road, CHELTENHAM, Glos., GL52 5AJ.

Telephone No. Cheltenham 21491 Ext. 2270.

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INTERNATIONAL AERADIO LTD **TELECOMMUNICATIONS** INSTRU

Applicants should possess a recognised qualification in telecommunications, e.g. City & Guilds, ONC, HNC with electronics. These are desirable gualifications but personnel with a sound basic and applied radio theory capability would be considered.

Ex-Services/Civilian personnel who have completed an instructional technique course and have had instructional experiences in the field of telecommunications would be preferred.

A liking and aptitude for this work is essential together with general experience of UHF, VHF and HF communications equipment providing fixed and mobile services. Applicants should also be familiar with Radio Navigational Aid equipment as installed at airports for use by aircraft and ATC personnel. It is inherent that applicants will have a knowledge of solid state techniques.

This is a permanent and pensionable position at our Radio Training School outside Southall which address is in easy access of surrounding areas. The post offers good career prospects. Starting salary will be in the region of £1500. Benefits also include membership of an excellent contributory pension and life assurance scheme and substantial rebates on holiday air fares, after a year's service

To apply for this position write to



Personnel Officer (Recruitment) INTERNATIONAL AERADIO LIMITED. HAYES ROAD, SOUTHALL, MIDDLESEX, Tel: 574 2411

THE HANNAH DAIRY RESEARCH INSTITUTE AYR **ELECTRONICS TECHNICIAN**

ASSISTANT EXPERIMENTAL/EXPERIMENTAL OFFICER required for duties in the Department of Physiology, including servicing of electronic and electrical equipment, construction of special instruments, devising and construction of electrophysiological apparatus. Previous experience in a physiological laboratory desirable, but not essential.

Qualifications: a degree, H.N.C., N.H.C., Grad. I.E.E.E. or equivalent in electronics engineering, applied science or applied physics. Salary: A.E.O. up to £1,208 in a scale to £1,454;

E.O. in a scale £1,190 to £2,006.

Further particulars may be obtained from the Secretary of the Institute, to whom applications, with the names of two referees, should be submitted by 17th January, 1970.

2734

2777

WE OFFER A YOUNG ENGINEER

the opportunity of working in an up-to-date tape recorder service department on Uher recording equipment.

The applicant should be familiar with the latest transistorised circuitry as well as being able to carry out mechanical work on such equipment.

We offer a good salary, non-contributory pension scheme, subsidised canteen facilities and some local transport.

If you are interested, please write giving brief details about your qualifications and experience to:

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 Heathkit, £35; 14 volumes "Radio and TV Servicing" until 1962. Offers considered.—H. F. ENGINEERING, 3 Willowbank, Sunbury-on-Thames, Middlesex. Telephone 83232. [2756]
 Mullard Double Beam Oscilloscope, L.101.
 Excellent general condition but may possibly need recalibrating. £80. (2) Marconi BFO, type TF195L.
 £30. (3) Avo model 7, £15. (4) Evershed Bridge Megger 500 volt, £20.—Burgess Lane & Co. Ltd., Thornton Ave. Chiswick, London W.4. [2638]

ROCRASS, etc., AR88, CR100, BRT400, G209, S640, etc., etc., in stock.—R. T. & I. Electronics, Ltd., Ashville Old Hall, Ashville Rd., London, E.11. Ley. [65]

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FOR hire CCTV equipment including cameras, monitors, video tape recorders and tape---any period. --Details from Zoom Television, Amersham 5001. [75]

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WANTED, all types of communications receivers and test equipment.—Details to R. T. & I. Electronics, Ltd., Ashville Old Hall, Ashville Rd., Lon-don, E.11. Ley. 4986.

don, E.11. Ley. 4986. [63] WANTED: Bruel & Kjaer automatic frequency response recorder type 3308 (2305/1022) 20 Hz-20 KHz (or earlier model) Details please to Mr. C. Heinlein, CTH ELECTRONICS, Hoddesdon, Herts. Phone Hoddesdon 64798. [2772] WANTED, televisions, tape recorders, radiograms, new valves, transistors, etc.—Stan Willetts, 37 High St., West Bromwich, Staffs. Tel. Wes. 0186. [72] WANTED: Wireless World 1938 Communication Receiver. Condition immaterial. Box WW 2773 Wireless World.

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WE buy new valves, transistors and clean new com-ponents, large or small quantities, all details, quotation by return.-Walton's Wireless Stores, 55 Worcester St., Wolverhampton.

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CAPACITY AVAILABLE A IRTRONICS, Ltd., for coil winding, assembly and wiring of electronic equipment, transistorised sub-unit sheet metal work.—Ja Walerand Rd., London, S.E.13. Tel. 01-852 1706. [61 ELECTRONIC and Electrical Manufacture and Assembly. Prototypes and short production runs. East Midlands Instrument Co. Ltd., Summergangs Lane, Gainsborough, Lincs. Tel. 3260. [88 M ETALWORK, all types cabinets, chassis, racks, to your own specification, capacity available for smail milling and capstan work up to lin bar.— PHILPOTT'S METALWORKS, Ltd., Chapman St., Loughborough. [17 CKILLED WIREMAN, own workshop, seeks out-work.

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KINGSTON-UPON-HULL Education Committee. College of Technology. Principal: E. Jones, M.Sc.,

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83

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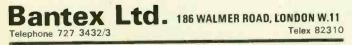
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2733



The photograph shows two boats of the Ford team in the 1969 Round Britain Power Boat Race. Both used Bantex aerials.



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INDEX TO ADVERTISERS

Appointments Vacant Advertisements appear on pages 95-109

A1 Factors 11	2	Hall Electric Ltd 9)
Acoustical Mfg. Co. Ltd.	4	Harris Electronics (London), Ltd 46	5
Adcola Products Ltd Cover i	iii	Harris, P 111	Į
Adler, B. & Sons (Radio) Ltd 3	8	Hart Electronics	
AEI Semiconductors Ltd 3	39	Harversons Surplus Co., Ltd 111	
Ahuja Radios 6	51	Hatfield Instruments, Ltd	
Amtronix	38	Hayden Laboratories 41	1
Anders Electronics, Ltd	4	Henry's Radio, Ltd 66, 67	1
A.P.A. Publishing (Catalogues) Ltd	23	Henson 112	2
	12	Hi-Fi Year Book/Radio Year Book Loose insert	
	24	Holford, F., Co., Ltd 112	
Avo Ltd	1	Howells Radio, Ltd. 58	3
	0		
	10 22	I.C.S., Ltd	
purnet	11	I.M.O. (Electronics), Ltd 35	
and the second s	44	Impectron Ltd	
and the second sec	58	Industrial Instruments, Ltd	
	92	Instructional Handbook Supplies 112	2
benney, it j	13		
Bi-Pak Semiconductors	72	Keyswitch Relays Ltd Cover in	
DI-I BR OCHHOOME BOTTO	93	Keytronics 110	
Black, J 111, 12		Kinver Electronics, Ltd. 110)
	15		
	40	Lasky's Radio, Ltd 75	5
bicite Big. Co., Etd.	42	Lawson Tubes 112	2
Bulgin, A. F., & Co., Ltd. Edit.		Ledon Instruments, Ltd 46	5
Buigin, A. F., & Co., Dia Internet in the		Levell Electronics, Ltd 18	3
Calan Electronics Ltd	51	Light Soldering Developments, Ltd 28	3
	40	Livingston Hire Ltd 70)
	6 3	Lloyd, J. J., Instruments Ltd 17	7
	62	London Central Radio Stores 111	L
	47	L.S.T. Components. 74	ł
Danavox (Gt. Britain) Ltd	19		
Daystrom, Ltd		Mainline Electronics, Ltd	
	92		
	33	Mills, W	
	70		
	84	M.R. Supplies, Ltd	
		Multicore Solders Ltd. Cover in	
ALLER ALLER CONTRACTOR	13	Myall, W. H	
	11	Myan, w. n	,
Electronic Brokers 64, 65, 1		Newmarket Transistors I td	2
Directionales (orb) ton);	69	Tewniarket Transistors Dear.	-
	78	Nombrex Ltd 42	6
	89	2	
Directive winder Directive and the second se	60	Omron Precision Controls	
	45	Osmabet Ltd	
E.M.I. Tape Ltd	52	Oxley Developments Co., Ltd	U
English Electric Valve Co., Ltd 3,	, 5		
Erie Electronics, Ltd.	16	Patrick & Kinnie	8
The H Treasure I ad	16	P.C. Radio, Ltd	
A BILLELI BILLELING,	46	Peak Sound (Harrow) Ltd 30	
	51	Pinnacle Electronics Ltd 25	
Futuristic Aids Ltd	47	Proops Bros. Ltd 7	
Garage Gifts Ltd 1	11	Pye Telecommunications, Ltd 2'	7
	26		
Contract mills, and mills	24	Quality Electronics Ltd 5	9
o o la	14	Quarndon Electronics Ltd 4	3
	39	Quartz Crystal Co. Ltd 9.	2

	PAGE
Hall Electric Ltd	9
Harris Electronics (London), Ltd	46
Harris, P	111
Hart Electronics	88
Harversons Surplus Co., Ltd.	111
Hatfield Instruments, Ltd	60
Hayden Laboratories	41
Henry's Radio, Ltd	
Henson	112
Hi-Fi Year Book/Radio Year BookLoose i	
	112
Holford, F., Co., Ltd.	_
Howells Radio, Ltd.	58
I.C.S., Ltd	, 112
I.M.O. (Electronics), Ltd.	35
Impectron Ltd.	32
Industrial Instruments, Ltd.	58
Instructional Handbook Supplies	112
Keyswitch Relays Ltd Co	ver ii
Keytronics	110
Kinver Electronics, Ltd.	110
Total Date Tak	75
Lasky's Radio, Ltd	
Lawson Tubes	112
Ledon Instruments, Ltd.	46
Levell Electronics, Ltd	18
Light Soldering Developments, Ltd	28
Livingston Hire Ltd	70
Lloyd, J. J., Instruments Ltd	17
London Central Radio Stores	111
L.S.T. Components.	74
Mainline Electronics, Ltd.	82
Marshall, A., & Son (London), Ltd	91
Mills, W	0, 81
Milward G. F.	83
M.R. Supplies, Ltd.	38
	1, 20
Multicore Solders Ltd Co	-
Myall, W. H.	60
Wiyan, w. H	00
Newmarket Transistors Ltd.	32
	42
Nombrex Ltd	44
Omron Precision Controls	35
Osmabet Ltd.	88
Oxley Developments Co., Ltd	60
Patrick & Kinnie	68
P.C. Radio, Ltd.	85
Peak Sound (Harrow) Ltd.	30
Peak Sound (Harrow) Ltd	25
Proops Bros. Ltd.	27
Pye Telecommunications, Ltd	21
Quality Electronics Ltd.	59
Quarndon Electronics Ltd	43
Quartz Crystal Co. Ltd.	92

	PAGE
Radio & TV Components, Ltd	87
Radio Components Specialists	113
Radio Exchange Co	92
Radiospares, Ltd.	42
Rank Wharfedale Ltd.	36
Rendar Instruments	60
Reslo Mikes	44
Roband Electronics Ltd.	34
R.R. Radio	112
R.S.C. Hi-Fi Centres, Ltd.	79
R.S.T. Valves	86
Salford Electrical Inst. Ltd.	47
Samsons (Electronics), Ltd.	80
S.E. Laboratories (Eng.) Ltd.	54
	2, 73
Servo & Electronic Sales, Ltd.	112
Simpson Taylor & Co. Ltd.	113
Sinclair Radionics, Ltd. 55, 5	
S.M.E. Ltd.	10
Smith, G. W., (Radio), Ltd	
Smith, H. L., Co. Ltd.	28
Smith, J., Ltd.	62
Smiths Radio Services (W'ton) Ltd	90
S.N.S. Communications Ltd	14
Special Products Ltd.	62
Specialist Switches, Ltd.	61
Stanford Electronics	114
Starman Tapes.	113
S.T.C. Electroniques	8
Sugden, J. E.	59
Sutton Electronics	112
Outton Estationation	
Techmation Ltd.	112
Teclare Ltd.	90
Tektronix Ltd.	12
Telequipment Ltd.	48
Telcon Metals Ltd.	26
Teonex, Ltd.	29
Thompson, A. J.	90
Thorn A.E.I. (Radio Valve & Tubes), Ltd	50
Trio Corporation	2
Trio Instruments Ltd.	59
T.R.S. Radio Components Specialists	84
T.A.O. Radio Componente op tomatica internet	
United-Carr Supplies, Ltd.	36
Universal.	92
Ontversal	14
Valradio, Ltd	8, 59
Vitality Bulbs, Ltd.	62
Vortexion, Ltd.	53
Conciding Enderson and the second sec	
Watts, Cecil E., Ltd.	58
Wayne Kerr, The, Co. Ltd.	31
Webber, R. A., Ltd.	59
West Hyde Developments, Ltd.	6
West London Direct Supplies	70
Wilkinsons, L. (Croydon), Ltd.	88
withinous, D. (Croydon), Det.	
Z. & I. Aero Services, Ltd	94
L. G. I. HELD DETTICES, EMANTHEIT	

Printed in Great Britain by Southwark Offset, 25 Lavington Street, London, S.E.1, and Published by the Proprietors, I.P.G. ELEGTRIGAL-ELECTRONIC PRESS, LFD., Dorset House, Stamford St., London, S.E.1, telephone 01-928 3333. Wireless World can be obtained abroad from the following: ADSTRALLA and New ZRALAND: Gordon & Gotch, Ltd. INDIA: A. H. Wheeler & Co. CANADA: The Wm. Dawson Subscription Service, Ltd.: Gordon & Gotch Ltd. Sourk Arxios: Central News Agency, Ltd.: William Dawson & Sons (S.A.), Ltd. UNITED STATES: Eastern News Co., 306 West 11th Street, New York 14. CONDITIONS OF SALE AND SUPPLY: This periodical is sold subject to the following conditions, namely that it shall not, without the written consent of the publichers first given, be lent, re-sold, hired out or otherwise disposed of in a mutilated condition or in any unauthorised cover by way of Trade or affixed to or as part of any publication or advertising, literary or pictorial matter whatsoever.

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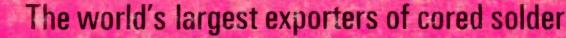
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