WirelessWorld Three Shillings

Review of digital microcircuits Logic display aid

May 1969

APR 1969

If you design and manufacture electronic equipment such as computers, radar systems, avionics etc., you will know all about Ferranti high precision transformers. You will know all about their quality and reliability, in short they are engineered by Ferranti standards. They have always been pretty popular—a bit too popular if anything, for we were continually pushed to meet delivery dates. Faced with this situation, we did the obvious thing expanded production. Not just a bit—but dramatically. A brand-new factory was raised at Dundee. New equipment installed—more production teams recruited and trained.

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C1179



C1148



C1150/1

New pulse tetrode for low power

radars added to EEV's range



	Service type	Anode dissipation max. (W)	Pulse A output ve power m (kW) D	Anode voltage	Pulse anode current max. (A)	Heater ratings		
Туре				max. D.C. (kV)		(∨)	(A)	Base
C1148	_	40	130	14.0	12	6.3	5.0	B5F
C1149/1	CV6131	60	330	20.0	18	26.0	2.15	B4A
C1150/1	CV427	60	205	17.5	15	26.0	2.15	B4A
C1166	_	60	205	17.5	15	6.3	9.0	B5F
C1179		18	65	8.0	9.0	6.3	2.8	B7A

Send for full data on the EEV range of pulse amplifier tetrodes



English Electric Valve Co Ltd Chelmsford Essex England Telephone: 61777

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 COMPANY
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WW13 AP 362

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BS390

BS310



Brief data on some of the many types available.

Send for this booklet giving full details of the complete range of EEV duplexer devices and waveguide switches.







BS824

BS458



BS452



BS460

		Í	Frequency range	Peak power
Product	Type No	Band	(MHz)	(kW)
Pre TR cells	BS834	-	2000-12000	2 <mark>500</mark>
	BS870	-	1240-1365	2500
TR cells	BS390	S	2925-3075	1250
	BS800	S	2840-3100	1250
	BS824*	S	2700-3100	250
	BS156	Х	9000-9600	200
	BS452	X	9310-9510	100
	BS810	X	9250-9550	75
	BS850	X	9300-9500	50
TB cells	BS310	X	9375	5-200
TR limiter cells	BS814	Х	9000-9700	200
	BS828	X	9325-9425	50
Solid state microwave switches	BS392	S	2925-3075	0.5
	BS460	X	8500-12000	0.5

*For protection of travelling waveguide amplifiers



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Frequency range	Power	Type of cell	
NAME		POSITION	
COMPANY			
ADDRESS			
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TELEPHONE NUMBER

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A7

EEV glass and ceramic hydrogen thyratrons are extensively used to provide more precise and efficient high speed switching. Here are some of the reasons why :

1 Their short anode delay time of between 20 and 120 nanoseconds depending on triggering method.

2 Low jitter generally of 1 to 2 nanoseconds but down to less than $\frac{1}{2}$ nanosecond depending on heater supply.

3 The negligible change in anode delay time typically only 10 nanoseconds over a long period of use.

4 A high peak inverse voltage capability of 20kV immediately following pulse.

5 The low trigger power required.

6 The wide operating voltage range of 1kV-120kV with four tubes.

7 The ability to control anode delay time and rise time of current, using reservoir.

8 The wide reservoir range for maintenance of gas pressure typically 4.5V to 5.7V.

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Medical linear accelerators with RF accelerating powers up to 15MW. Particle linear accelerators with RF accelerating powers up to 50MW. They may also be used in first-stage particle beam choppers. Particle beam benders where a network of stored energy needs to be discharged into a deflection coil or other device somewhere on the accelerating ing.

Spark chambers For pulsing light shutters such as Kerr or Pockel cells.

Electronic crowbars and energy diverters

EEV thyratronsfor better high speed switching



Туре	Peak power output max (MW)	Heating Factor (V.A.p.p.s.)	Peak forward voltage max (kV)	Peak anode current max (A)	Mean anode current max (A)
CX1154	50.0	30 x 10°	40	2500	3.0
CX1157	3.5	7 x 10 ⁹	20	350	0.35
CX1168	100.0	70 x 10 ⁹	80	2500	2.5
CX1171	150	70 x 10 ⁹	120	2500	2.5
CX1174	120	60 x 10 ⁹	40	6000	6.0
CX1175	200	140 x 10°	80	5000	6.0
CX1180	12.5	9 x 10 ⁹	25	1000	1.25

Send for full details of the complete range of EEV thyratrons.

Brief data on some of the ceramic types

available.



English Electric Valve Co Ltd

Chelmsford Essex England Telephone: 61777 Telex: 99103 Grams: Enelectico Chelmsford



Please send me full data on your complete range of glass and ceramic hydrogen thyratrons I am particularly interested in using a thyratron with the POSITION NAME following parameters: COMPANY Application ADDRESS Peak power output Peak forward voltage WW15 TELEPHONE NUMBER **EXTENSION** Peak anode current AP 359

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Be safe...use EEV magnetrons in your marine radar

	Туре	Frequency Range (MHz)	Power (kW) (Typical Operation)	Equivalents (not complete)
Brief data on some of the many types available. The	M5063	3025-3075	50	2J70B
	2J42	9345-9475	8	ME1101, CV3676, MAG3, M526
S-Band and X-Band	BM1002	9415-9465	21	JP9-15B
ypes from 3-80kW.	M513B	9345-9405	22	JP9-15, YJ1110
	M515	9380-9440	25	YJ1120
	M597	9380-9440	10	
	M598B	9380-9440	22	
	599A/B	9415-9475	3	JP9-2.5D, JP9-2.5E, 7028
	M5022	9415-9475	30	YJ1121
	M5031	9345-9405	9	
	M5043	9380-9440	5.8	
	M5039	9345-9405	22.5	



1.100

M515

M599A/B



M513B

English Electric Valve Co Ltd

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	Chelmsford Essex England Telephone: 61772
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Please send me full data on your range of marine magnetrons. I am particularly interested in using a marine magnetron with the following parameters. Peak Output Pulse **Pulse Repetition** Frequency Range (MHz) Power (kW) Length (µs) Rate (p.p.s.) NAME POSITION COMPANY ADDRESS

TELEPHONE NUMBER

EXTENSION

WW16

WW—014 FOR FURTHER DETAILS

Exploring the Hewlett-Packard Universe of Electronics Instrumentation

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1 Two oscillators from among 17 2 Plug-in scope system 3 Low-priced digital voltmeter 4 Universal counter 5 Hewlett-Packard Journal



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Now there are 17 different oscillators, including two new ones we'd like you meet. Both feature 0.5% (0.05 dB) flatness, FET's in the bridge for improved stability, < 0.1% (-60 dB) distortion, and balanced output.

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2 A scope system that's big in versatility and small in size

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A10

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E 26-UK

3 A digital voltmeter for the fair sex?



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100 ns to 108 s.

just about all the measurements for which electric counters are used. The 5325B thus measures frequency, period, multiple period averages, ratio, multi-ratios and simple or complex time intervals from

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5 An insider's view of **R&D** at Hewlett-Packard

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4 Why we call it a universal counter



We call the hp 5325B a universal counter because it has built-in capability to perform

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WW-034 FOR FURTHER DETAILS

Wireless World, May 1969

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WW-035 FOR FURTHER DETAILS

G 18 watts. Ideal for miniature work on production lines. Interchangeable spare bits, 3/32", 1/8", 3/16", and 1/4". For 240, 220 or 110 volts, 32/6. E 20 watts. Fitted with 1/4" bit. Interchangeable spare bits 3/32", 1/8 3/16". For 240. 220. 110 or 24 volts.from 35/-ES 25 watts. Fitted with 1/8" bit. Interchangeable bits 3/32", 3/16" and 1/4". Ideal for high speed production lines. For 240, 220, 110, 24 or 12 volts. from 35/-

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WW----036 FOR FURTHER DETAILS



WW-037 FOR FURTHER DETAILS



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Wireless World, May 1969

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A38

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audio tone burst generator



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start, it combines accuracy and

testing of equipment where the source impedance is high. You can use them with your old test meter, but, if you want the meter to last as long as the probe, you'll have to get a Selectest.

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WW-057 FOR FURTHER DETAILS



offer a selection from our standard range

-15" (3.81 mm) TAPE HEADS (CASSETTE)								
TRACK CONFIG.	TRACK SERIES APPLICATIO		INDUCTA R/P	NCE mH	TYPE No.			
W		Rec/Play Rec/Play Erase	75 65	- 0.65	W12RP56* W12RP65 W12E340*			
1/2	AW	Rec/Play Rec/Play/Erase Erase	110 110	- 1·5 1·5	AW12RP90 AW12RPE91 AW12E360			
2/2	AW	Rec/Play Rec/Play/Erase Erase	70 70	1.5 1.5	AW22RP103 AW22RPE104 AW22E366			
2/4 AW Rec/Pla Rec/Play/E		Rec/Play Rec/Play/Erase	110 70	- 1·2	AW24RP92 AW24RPE93			
4/4	AW	Rec/Play Rec/Play/Erase	30 30	- 1·0	AW44RP94 AW44RPE95			

1/4" (6:35 mm) TAPE HEADS							
TRACK CONFIG.	SERIES	APPLICATION	INDUCTA Parallel	NCE mH Series	TYPE No.		
*		Rec/Play	1.25 7.5 27.5 162.5	5.0 30.0 110.0 650.0	X11RP26 X11RP70 X11RP25* X11RP20		
1/1	x	Record	1.0 6.0 20.0	4.0 24.0 80.0	X11R71 X11R72 X11R27		
		Erase	тХт	0.55 2.5 8.5	X11E350 X11E316* X11E351		
		Rec/Play		5.0 30.0 70.0 110.0 650.0 1000.0	X12RP23 X12RP42 X12RP41 X12RP15* X12RP14* X12RP14*		
	x /	Record	1.0 6.0 20.0	4·0 24·0 80·0	X12R73 X12R74 X12R16*		
	/ 2	Rec/Play/Erase	R/P- E -	200-0 0-5	X12RPE107		
1/2		Erase		0.3 1.5 2.5 5.0	X12E352 X12E353 X12E317* X12E334		
		Rec/Play	1.25 7.5 30.0 140.0	5.0 30.0 120.0 560.0	BX12RP75 BX12RP76 BX12RP77* BX12RP63*		
	BX	Erase	- - - 3 m/	0,4 1.75 2.5 6.0 D.C.	BX12E356 BX12E357 BX12E343* BX12E358 BX12E344*		



AW12RP

Width '44". Height '32". Length :65". Head for '15" wide tape for Cassette Application. The mounting plate is an integral part of the head providing simple azimuth adjustment. Built in tape guides are another feature of this head. Various special Record/Play and Erase heads are made to customers own regulrements.

W10RP

Various special versions of narrow track and protruding pole heads are available in the W series, for Clne and Dictating machine applications etc. Example shown is for 8 mm. Cine and Is a Record/Play head having protruding pole of .02" track width.





X10RPE

An example of narrow track Record/Play/erase heads for Dictating machine application—Record/Play section .010", Erase section .014". Can be supplied in self oscillatory version if required. Erase section gives self biasing effect. Deep drawn mumetal case $\frac{1}{2}$ " × $\frac{1}{2}$ " × .55" deep provides adequate shielding.

CX28RP

 $\frac{1}{2}$ " × $\frac{1}{2}$ " × -55" deep drawn mumetal case. This type of head is for 8 tracks on $\frac{1}{2}$ " tape. Movement of the head across the tape is used to achieve this result. Cross talk figure better than ---70 dB's is obtained with this arrangement. Can be supplied in variety of inductances to customers own requirements.





X24RP

Standard "X" series Head in deep drawn mumetal case $\frac{1}{2}$ " $\times \frac{1}{2}$ " $\times \frac{1}{55}$ ". For use in high quality tape recorders and available in a wide range of inductances. Excellent high frequency performance, efficient screening, very low cross talk. Over a million and a half of these heads have been manufactured.

X12RPE107

 $\frac{1}{2}^{\circ} \times \frac{1}{2}^{\circ} \times \cdot 55^{\circ}$ deep. 1/2 track combination head for applications such as telephone answering machines. Particularly useful where head space is limited. Overcomes any problems of alignment of Erase to R/P tracks. Deep drawn mumetal case offers perfect screening. Other impedances are available to customers requirements.





BX Series

A new series of Record/Play head in $\frac{1}{2}$ " $\times \frac{1}{2}$ " $\times \frac{1}{55}$ " deep mumetal case. This type has been developed particularly as a replacement for $\frac{1}{2}$ " square heads commonly used on various tape recorders of the more mass produced variety. Lower priced than the X series. An Erase head of similar size is available in deep drawn brass case.

WW-058 FOR FURTHER DETAILS

DR Series

•437" × •437" × •625" depth in die-cast body with nlckel silver front provided with fixing holes for PK screws. The R/P head has internal mumetal screen. The Erase head is double gap, double field variety. The cheapest series of heads made by Marriott Magnetics. An Improved design of the first mass produced head to be made in the world in 1957 by our company.





R Series

Size of the front '437" × '437" with 3" body diameter and 3" long. This head is available in a wide range of Record/ Play Impedances electrically similar to DR type. Head body is made in brass and has internal mumetal screens. Offers special advantages for easy mounting and azimuth adjustment. Erase heads, electrically as DR series, are available.

A12RP

Combination Record/Playback/Erase head in deep drawn mumetal case $\frac{1}{2}$ × $\frac{1}{2}$ × \cdot 55" deep. Particularly useful where head space is limited. Combination arrangement ensures correct alignment of Erase tracks to Record/Play tracks. Available in various impedances other than those

X24E

a" diameter, §" long. Standards available are for 1/2 track configuration but many special versions can be made such as narrow track, protruding poles, cut away edge for cine use etc. The round body makes for easy azimuth adjustment and takes up a minimum of space. The head incorporates an internal screen and flying leads. Special versions are available with ferrite poles for drum applications.

X24RPE

listed.





This 2/4 stereo head size $\frac{1}{2}$ × $\frac{1}{2}$ × $\frac{1}{2}$ × $\frac{55}{2}$ deep is a companion to all X series Record/Playback heads and is available in a wide range of impedances. It is a highly efficient double gap erase head which leaves the tape with a very low noise level after erasure.



CX88RP

This remarkable head built into a standard deep drawn mumetal case $\frac{1}{2}$ " $\times \frac{1}{2}$ " $\times \cdot 55$ " deep gives 8 track Record/ Playback facilities on $\frac{1}{4}$ " tape. Marriott Magnetics are once again the first Company in the world to quantity produce such heads and at prices not previously thought possible. Possibilities now created for digital data collection etc., utilising $\frac{1}{4}$ " tape transport mechanisms. Already being fitted to various Cassette systems in the U.K. The 4/8 version offers low enough crosstalk figures for audio applications.

T10RPE

Dimensions 5" by 3" by 55" length. Example of a protruding pole type of head with special narrow track developed for Dictating Machines. Example shown is a Record/ Playback/Erase head of self oscillatory variety and each section incorporates a transformer coupling so that DC can be passed direct through the head. Mumetal shielded case and fully screened leads.



A new series of narrow width Erase heads (not illustrated) will shortly be available for mounting alongside standard X series Record/Replay heads. Width of these heads is only .16"

*Types produced in large quantities and usually held in stock.

1/4 (6.35 mm) TAPE HEADS contd.							
TRACK CONFIG.	SERIES	APPLICATION	INDUCTA Rec/Play	NCE mH Erase	TYPE No.		
	DR	Rec/Play	5.0 30.0 70.0 110.0 250.0 650.0		DR12RP50 DR12RP43 DR12RP45 DR12RP45 DR12RP33 DR12RP35 DR12RP31		
1/0		Erase	1111	0·2 1·5 2·5 3·0 7·0	DR12E354 DR12E301 DR12E305 DR12E355* DR12E397*		
1/2	R	AVA	ILABLE IN	ALL DR TY	PES		
	А	Rec/Play	2·25 5·0 30·0 70·0 110·0	1111	A12RP37 A12RP98 A12RP99 A12RP99 A12RP100 A12RP101		
		Erase		0·15 0·7 1·25	A12E364 A12E365 A12E330*		
	x	Rec/Play	5.0 30.0 110.0 650.0	1111	X22RP80 X22RP81 X22RP46* X22RP47*		
2/2		Record	4.0 24.0 80.0	1 - 1	X22R82 X22R83 X22R48*		
- 1		Erase	1111	-56 1-4 2-5 5-0	X22E341 X22E359 X22E333* X22E336		
1/4	X	X Rec/Play/Erase		1.0	X14RPE55		
		Rec/Play	4.0 30.0 70.0 110.0 650.0	-	X24RP24 X24RP30* X24RP36* X24RP36* X24RP18* X24RP17*		
2/4	x	Record	4.0 15.0 30.0 100.0 500.0	1 1 1 1	X24R84 X24R44 X24R85 X24R49* X24R28		
		Rec/Play/Erase	110.0	1.0	X24RPE108		
		Erase	1111	0.5 0.2 1.0 5.0	X24E332A X24E342* X24E311* X24E335		
4/4		Rec/Play Rec/Play/Erase	110·0 110·0	1.0	X44RP105 X44RPE106		
1/8 2/8 4/8 8/8	сх	Rec/Play	30.0 30.0 30.0 30.0 30.0	1111	CX18RP86 CX28RP87 CX48RP88 CX88RP89		
	т	Rec/Play Rec/Play/Erase Playback	85·0 40·0 100·0	Self Osc.	T10RP53 T10RPE54 T10P60		
	X	Playback Playback	250-0 5-0	-	X10P57 X10P58		

1/2" (12.7 mm) and 1" (26.4 mm) TAPE HEADS

Marriot Magnetics have hitherto concentrated principally on special versions of such heads to customers own requirements. We are about to embark on producing a standard range of heads for $\frac{1}{2}$ and 1" tape. We would like to hear of your own requirements in this field. Our prices will be considerably lower than any other to be found in the world, in view of the quantities we intend to produce.

IN ADDITION TO THE ABOVE-MARRIOTT MAGNETICS MANUFACTURE A LARGE VARIETY OF HEADS FOR SPECIAL APPLICATIONS SUCH AS CINE DRUM, VIDEO ETC. TOGETHER WITH A WIDE RANGE TO CUSTOMERS REQUIREMENTS.



WW-059 FOR FURTHER DETAILS

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MU.7522	3.75/15*	1-3, 2-4	100K.	6-8	82:1/164:1	Low Z. Mic/Grid
MU.7523	75/300*	1-3, 2-4	600 (C.T.)	6-7-8	1.41:1/2.82:1	Line/Line
MU.7524	150/600*	1-3, 2-4	600 (C.T.)	6-7-8	1:1/2:1	Mixing :Bal./Unbal.
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MU.7526	600 (C.T.)	6-7-8	2.5k/10k.*	1-3, 2-4	2.04:1/4.08:1	Line/Grid
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MU.7528	7 <mark>·5/30</mark> *	1 <mark>-3, 2</mark> -4	600 (C.T.)	6-7-8	4·47 :1/8·94 :1	Low Z. Mic./Line
MU.7529	50/200*	1-3, 2-4	600 (C.T.)	6-7-8	1· 73 :1/3·46 :1	Mic. or Line/Line
MU.7530	10K. (C.T.)	6-7-8	<mark>10К.</mark>	1-4	1 (C.T.) :1	600 Line Bridging
MU.7532	7.5/30*	1 <mark>-3, 2-4</mark>	<mark>100к.</mark>	6-8	58:1/ <mark>116:1</mark>	Low Z. Mic./Grid
MU.7534	50/200*	1-3 <mark>, 2-</mark> 4	100K.	6-8	22.4:1/44.8:1	Mic. or Line/Grid
MU.7528 MU.7529 MU.7530 MU.7532 MU.7534	7·5/30* 50/200* 10K. (C.T.) 7·5/30* 50/200*	1-3, 2-4 1-3, 2-4 1-3, 2-4 6-7-8 1-3, 2-4 1-3, 2-4	100K. 600 (C.T.) 600 (C.T.) 10K. 100K. 100K.	6-8 6-7-8 6-7-8 1-4 6-8 6-8	13 : 1/26 : 1 4·47 : 1/8·94 : 1 1 ·73 : 1/3·46 : 1 1 (C.T.) : 1 58 : 1/116 : 1 22·4 : 1/44·8 : 1	Line/Grid Low Z. Mic./Line Mic. or Line/Line 600 Line Bridging Low Z. Mic./Grid Mic. or Line/Grid

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Where the standard Volume Indicator measures only the top 23 dB of signal level logarithmically, this new Program Monitor displays information from +3 to -57 dB on a single linear, decibel scale, thus permitting accurate reading of low level audio material as well as line noise during program pauses. The 0 dB reference settings are adjustable from +18 to -22 dBm.

The 600 is also equipped with a separate DC output for graphic logging over the full 60 dB range or to drive a second meter for remote monitoring. It is also available in a standard 19-inch mounting rack from which it can be easily removed for portable use.

While not intended as a replacement for the standard Volume Indicator, the 600's meter ballistics are such that its readings are compatible with VU indications. It's a practical program monitor as well as a valuable measuring tool.

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Wireless World

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Dur cover picture this month is the Venn liagram for the sum output of a full adder, produced on the screen of an oscilloscope, by the Logic Display Aid described in the series of articles beginning in this issue. The oscilloscope was deliberately defocused to produce the extra-large dots.

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How Mullard developed the valves for today's hybrid TV sets

During the earliest stages of semiconductor development, Mullard recognised that the all-valved television receiver, whilst giving reliable performance and economic set design, could with advantage incorporate semiconductor devices in place of those valves used in the low signal handling stages. Consequently, Mullard pioneered the design of hybrid television and were the first to offer a complete set of valves for the purpose. Today, we are Europe's major supplier of complete ranges of valves both for colour and monochrome sets.

Each valve provides a low cost solution to the design problems found in the critical high power deflection and output stages of television receivers.

Purpose designed Each valve performs a specific function in parttransistorised receivers. But, before developing these valves, Mullard applications laboratories had to solve the complex problems of matching the optimum specification for each individual valve stage in a hybrid circuit layout. Nothing was left out—chassis tolerances, component stability, reliability, life performance, supply variation—all were investigated and specified.

Consistent quality All the plant, equipment and component parts for manufacturing valves were designed and built by ourselves at our Blackburn factory. In fact, our reputation for consistent product quality is a direct result of this 'do-it-yourself' policy, coupled with quality control that starts at the raw-material stage. We even produce our own grid wires from tungsten powder. And we process the critical cathode-emission coating, using barium, strontium and calcium nitrates that comply with our very tight specifications. The same tight control is exercised right down the production line, offering setmakers top-quality, reliable products at an economic price.

Continuous improvements Just because we produced the best possible valves to start with, it doesn't mean that development is forgotten. Whenever a new material or a new method of production arises from research studies or factory development projects, we investigate to see if it offers an improvement.

Complete data for set designers Mullard valves are supported by comprehensive data in the form that designers appreciate. For example, the data for deflection valves includes

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design charts which make full allowance for valve and component tolerances, for performance changes with valve life and for mains voltage variations.

Sales Setmakers appreciate the overall quality and economy of Mullard valves for hybrid TV, because most new television sets, both colour and monochrome, in the UK, now have them fitted as standard. Overseas customers are also specifying Mullard valves in large quantities.

Worth it? Right from the beginning we've had everything under our control, so that we can be sure the product will give consistent service. This also enables us to relate quality with the best possible price. Something which applies across the very wide Mullard component range. Our components find applications as unexpected as Astronomy and Zoology. And because of the many and unusual applications for our components, we have experience in many technologies. Experience our customers now take for granted.

Mullard components for consumer electronics

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Wireless World

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Nationalized in all but name

A recent announcement from one of this country's major electronics companies stated that the product about which they were writing was developed as a *private venture*. It was obvious that the company, or at least its P.R.O., considered that this fact was worth shouting about, presumably because it was the exception to the norm.

Hard on the heels of this announcement came the news from the National Research Development Corporation that, at the request of the Ministry of Technology, the Corporation is to invest nearly £5m in "the development and manufacture of microelectronic devices by Ferranti Ltd, Marconi-Elliott Microelectronics Ltd, and the Plessey Company Ltd. . . . The Corporation expects to recover its investments with interest by 1980 through a levy on the companies sales of silicon integrated circuits".

Such announcements as these only underline the fact that the electronics industry in general is becoming more and more dependent directly or indirectly on Government support. In addition to such direct financial assistance as that cited above—and there are many others—the Ministry of Technology is the country's biggest customer for electronic equipment.

In our report, elsewhere in the issue, on the Paris components show reference is made to the possibility of the U.S.S.R. becoming a major contender in the international microelectronics race. If this does happen, as is highly probable, then in the face of such national competition and that already facing us from across the Atlantic (from companies bolstered up by military and space spending) it may well be inevitable for our own industry to have substantial Government backing if it is to survive. On the other hand there is a lurking danger in the security enjoyed under a beneficent Government. It can, of course, have a stultifying effect on initiative; indeed the N.R.D.C. microelectronics announcement mentioned above stresses that "an important feature of the arrangements has been the establishment of a collaborative agreement between companies themselves. Each company has undertaken to disclose broad details of its own research and development to its collaborators". It would therefore appear that it is no longer necessary for any one of the three companies to try to get the edge on the others by research, ingenuity, skill, or what have you. It can be, and is of course, argued that duplicated, or in this case triplicated, research and development is wasteful. However, a much more serious aspect of corporate research was stressed by Commander H. Pasley-Tyler, retiring president of the Electronic Engineering Association, speaking at the annual luncheon of the E.E.A. on March 26th. That is, that too high a percentage of our research and development resources are locked up in laboratories which serve a number of official and semi-official organizations for the fulfilment of their own narrow purposes without any relevance to the wider possibilities of international sales.

Our industry (and for that matter much of the technology on which it is based) was built on Government spending during World War II and over the years it has tended to rely, perhaps a little too heavily, on this source of income. Military spending has, however, been drastically cut and the industry is feeling the pinch. This may be a good thing as it could stir us to seek new pastures, but it would appear that the industry is still relying on the umbilical cord to the Mintech remaining intact. Cmdr. Pasley-Tyler criticized in relation to the curtailment of military spending, the "confident belief that the civil side of the Ministry of Technology will somehow look after our technological future".

With the growing number of mergers, resulting in a few large units making up the bulk of the industry, which is becoming increasingly dependent upon Government support, one is tempted to ask how long it will be before there is a move to make unofficial nationalization official.

Editor-in-chief: W. T. COCKING, F.I.E.E.

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Assistant Editors: -B. S. CRANK -J. H. WEADEN

Editorial Assistant

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-Drawing Office: H. J. COOKE

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Wireless World Logic Display Aid

1: Introduction

designed by B. S. Crank*

The current methods of teaching logic design are lacking in some respects as there is a gulf between the practical and theoretical aspects of tuition.

The normal procedure is to introduce the student to a "logic tutor" after some initial training has been done in the elements of Boolean algebra. Representative circuits are "patched-up" and the outputs are interpreted from lights, meters or in fact from anything that is capable of a two-state indication.

Excellent ways of representing logic functions graphically already exist, in the form of Venn diagrams, Karnaugh maps and Truth tables, and these have the advantage of presenting the abstract functions in ways that can be easily visualized. This latter point is important as the student who can visualize logic combinations will soon reach a much deeper understanding of the underlying fundamentals. It is almost certain that during the theoretical instruction at least one of the above methods of representation, more likely all three, was employed.

As soon as practical work is started the method of display is reduced to flashing lights and other two-state indicators. These types of indicators have to be interpreted by the student and, since they are not graphically representative it is most unlikely that he will be able to visualize the function being demonstrated and, therefore, he may miss the point.

The Wireless World Logic Display Aid[†] combats this problem by producing, on a standard oscilloscope, the Venn diagram, Karnaugh map or the Truth table of any gate or logic circuit that is connected to the display aid.

For instance, if a binary adder circuit were connected to the display aid, and if the instrument were switched to the Truth table mode, the rules of binary addition, which is the adder Truth table, would be displayed on the oscilloscope. In other words the hardware itself produces exactly the same display as was used by the instructor on the blackboard during the theoretical sessions and, as a result, the problems confronting the instructor and the student are much reduced.

Some applications of the instrument are listed below: others will suggest themselves to readers as they become more acquainted with the device.

Applications

- 1. Teaching Boolean algebra.
- 2. Introducing and explaining the properties of Venn diagrams, Truth tables and Karnaugh maps.
- 3. Demonstrating the basic logic functions: AND, OR, NAND, NOR, etc.
- 4. Showing how gates can be combined to satisfy complex Boolean equations.

*Assistant editor Wireless World.

† Provisional patent specification No. 14062/1969.

- 5. Explaining the difference between positive and negative logic and demonstrating the relationships between AND OR- NAND- NOR and the effects on the hardware of a change in logic convention.
- 6. Explaining what minimization is and how it is accomplished.
- 7. As an aid to minimization of logic systems.
- 8. Shows how binary arithmetic can be performed with logic.
- 9. As an aid to teaching modern mathematics.
- 10. Quickly tests the results of practical work carried out by students.
- 11. Can be used as check-out equipment on production lines manufacturing logic sub-assemblies.
- 12. As a bench test equipment for rapid fault finding on logic assemblies.
- 13. Can be built into equipment for monitoring purposes.

By way of an experiment the writer introduced the instrument to his seven- and five-year old daughters who had no prior knowledge of the Venn diagram. The



The completed prototype which incorporates all the extra facilities mentioned on page 198. The four sets of function control switches, one set for each display area, can be clearly seen. Two input sockets for the external logic circuits are provided. The two push-buttons on the right of each set select either external circuit 1 or 2 for the appropriate display area; pressing both of these buttons results in the difference between the two circuits being displayed. The set of terminals on the left are for the variable outputs (A, B, C, D), the set in the centre are for the outputs of the external logic circuits (Z_1, Z_2) and the terminals on the right are power supplies for external logic circuits and for the extra variables (E, F) when the instrument is used in the 6-variable Karnaugh map mode.

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'lesson" was treated as a game. The square on the scilloscope screen which contains the three interlocking ircles of the Venn diagram *(universe)* was called a arden. The three circles representing the variables vere called the areas where plum, apple and orange trees row. Within five minutes both children could recognize ll the individual areas of the Venn diagram. Encouraged, he writer introduced the children to the Karnaugh map nd Truth table in turn. In a very short time they could nterpret the meanings of both.

The form which the display takes can be clearly seen n the accompanying photographs which show the instrutent operating in its various modes.

Before commencing the description of the instrument few words about the form of the series of articles in *Vireless' World* would not go amiss. The instrument an be easily divided into a number of small sections, ach section carrying out a particular sub-function. One ub-function at a time will be described in its entirety. That is, first the theory behind the sub-function will be iscussed followed by a description of the circuit, contruction and testing. This means that construction work ould, if desired, be started long before the series of rticles is finished. However, the instrument is a fairly omplicated one and the inexperienced constructor is dvised to wait until part five of the series is published efore deciding if he is competent to start the project.

This first article is introductory; the overall functionig of the instrument will be discussed and some of the elevant basic theory will briefly reviewed. Subsequent rticles will give detailed constructional information and ill also describe some accessory units that may be used ith the instrument.

The instrument is constructed using 42 integrated -ircuits. In general components are mounted on small -tug-in cards. The integrated circuits are from the erranti series 300 Micronor II range of diode-transistor ogic units which are in 14-lead plastic dual-in-line -tckages. These will be described in more detail later. I the design cost was considered to be secondary to orsatility, reliability and ease of operation.

eneral description

he block diagram of the complete instrument is shown in ig. 1. Although not immediately obvious from this drawig, the instrument is divided into two distinct sections: rcuits for deriving oscilloscope scan voltages and logic rcuits for obtaining the video signals. The external gic circuit connected to the instrument by the user -tercises a significant amount of control over the internal gic section which produces the video signal.

The scan voltages and the video signals are derived om two counters called the X and Y counters and it is —ith these that the description will begin, Each counter consists of four bistables connected to count in natural binary as shown in table one. The table corresponds to the output of the Y counter, the bistables of which are labelled A, B, C and D. The bistables in the X counter are labelled E, F, G, and H and follow the same counting sequence as the Y counter. A and E are the least significant. In the table, 1 corresponds to a positive voltage and 0 to a voltage very near to earth potential. The Y counter is driven by a multivibrator at about 20kHz. The output of the Y counter forms the input to the X counter.

D	С	В	A				
0	0	0	0				
0	0	0	1				
0	0	1	0				
0	0	1	1				
0	1	0	0				
0	1	0	1				
0	1	1	0				
0	1	1	1				
1	0	0	0				
1	0	0	1				
1	0	1	0				
1	0	1	1				
1	1	0	0				
1	1	0	1				
1	1	1	0				
1	1	1	1				
0	0	0	0				
etc.							

In a television set the spot is moved across the face of the screen by two sawtooth voltages. A similar method of deflection is used in this unit, the only difference being that the scan voltages are staircase waveforms instead of sawteeth.

The staircase scan waveforms are produced by the two digital-to-analogue converters (dians), one of which is connected to each counter. The voltage output of each dian is proportional to the numerical contents of the counter to which it is connected. Taking the Y counter as an example; at each input pulse the contents of the counter increase by one and the output of the dian alters by one unit. After the counter has received 16 input pulses the next pulse will return the counter to the "all-zero" position; this transition corresponds to flyback. The staircase produced by the Y dian is shown in Fig.2. The Y deflection voltage is precisely 16 times the frequency of the X deflection voltage because each counter divides by 16. The outputs of the dians drive the c.r.t. deflection plates.



ig. 1. A block diagram of the complete instrument.

It is arranged that as the contents of the Y counter increase, the output of the Y dian goes negative moving the spot down the face of the screen. With the X waveform the reverse is true; as the contents of the counter increase the putput of the X dian goes positive, moving the spot to the right.

Imagine that only the Y output voltage is connected to the oscilloscope. The spot will be moved down the face of the screen in 16 discrete steps so that a vertical column of 16 dots will be traced on the tube face.

With both deflection voltages (X and Y) connected to the oscilloscope, when the spot is at the bottom of the screen the next pulse from the multivibrator will cause the contents of the Y counter to fall to zero and the output of the Y counter will increase the contents of the X counter by one. As a result the spot will fly back to the top of the screen in a position slightly to the right of the column of dots it has just traced out. Column after column of dots will be traced until both counters are "full". The next input pulse causes both counters to return to zero and the spot to fly back to the top left-hand corner of the screen.

The "raster" produced by these circuits will consist of a square of dots with 16 rows and 16 columns, 256 dots in all. It is on this matrix of dots that all patterns are based. The matrix-raster is shown in Fig. 3.

The next section of the instrument, the circuits for forming the video signal will be discussed in principle

C

Photographs of the oscilloscope screen showing the display in operation. It is stressed that these patterns were produced by the prototype which has all the additions mentioned on page 198. The basic Logic Display Aid will only produce one of the four maps shown in each photograph at any one time. In photograph (a) the instrument was connected to a full binary adder and the SUM output is displayed. The whole left-hand part of the photograph is the Truth table for the SUM output of the binary adder; the first column is A, the second B, the third C and the fourth the result or SUM. The top right-hand display (still photograph (a)) is the Karnaugh map for the function and below it is the appropriate Venn diagram. In photograph (b) the external logic circuit was an AND gate connected to the $\overline{A}, \overline{B}$ and \overline{C} terminals, the positions of the Venn diagram and Karnaugh map are reversed when compared to (a).

Photographs (c) shows how two different circuit functions can be displayed simultaneously, the Venn diagram and Karnaugh map for A and \overline{A} being displayed. For photograph (c) two binary adders, one of which was not functioning correctly, were connected to the display aid. The left side of the picture shows the Venn diagram for each of the two adders; the top right display is the Venn diagram, and the bottom right is the Karnaugh map, for the difference between the two adders showing that the term $\overline{A} \overline{B} C$ is missing in one adder.

A

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only at this stage, a fuller description being given later.

If a binary-to-decimal converter was connected to the Y counter it would have 16 outputs to cope with all possible conditions of the counter. In such a converter one of these outputs would correspond to each state of the counter. Therefore each output of the converter would correspond to a particular row in the matrix raster. This point is illustrated in Fig. 4. The same sort of converter could be connected to the X counter, only this time the outputs would correspond to particular columns in the matrix.

If one of the outputs of the Y converter is AND gated with one of the outputs of the X converter the output of the AND gate will be "up" (at a positive voltage) only when the spot is at the intersection of the row and the column chosen. In other words a particular dot within the matrix can be selected. This point is illustrated in Fig. 5. If the output of the AND gate is presented to the Z terminal (intensity modulation) of the oscilloscope, and the brightness control is correctly adjusted, only the selected dot, in its correct position, will be visible on the tube face. It is easy to see how the process could be extended to select a number of dots to form any particular pattern.

In the instrument, although the above principle is em--ployed, binary to decimal converters are not used because -such an approach would be very expensive. The code converters shown in the block diagram of Fig. 1 modify the binary output of the converters in such a way as to



Fig. 2. The staircase waveform produced by the Y digital-toinalogue converter.



make the subsequent decoding of the various patterns an easier task. We will break from the description of the instrument for a short while and briefly review some important points of theory.

Boolean algebra

It is beyond the scope of this article to give any comprehensive explanation of this subject. Much has been written in these pages in the past and there are numerous books available. However a few of the basic rules are stated below:

$$A + 0 = A \quad A.A = A$$

$$A + 1 = 1 \quad A + \overline{A} = 1$$

$$A 0 = 0 \quad A \overline{A} = 0$$

$$A + A = A \quad A 1 = A$$

$$\overline{A \overline{B} \overline{C}} = \overline{A} + \overline{B} + \overline{C}$$

$$\overline{ABC} = \overline{A} + \overline{B} + \overline{C}$$

$$(De \text{ Morgan's Theorem})$$

$$AB\overline{C} + ABC = AB(C + \overline{C}) = AB$$

$$(\overline{A} + \overline{B}) (A + B) = \overline{A}B + A\overline{B}$$

$$\overline{\overline{A}} = A$$

Venn diagrams

The Venn diagram is a method of graphically representing a Boolean function. It consists of three interlocking circles within a square as shown in Fig. 6. Each circle represents one variable. The circles are normally labelled A, B and C. Everything that is outside circle A represents \overline{A} so the area outside all three circles is \overline{ABC} . The square is divided into eight separate areas by the circles as there are eight possible ways of com-



Fig. 4. Shows how a binary-to-decimal converter could be used to address any particular row of dots in the matrix-raster.



Fig. 5. Two binary-to-decimal converters in conjunction with an AND gate can select any one do't within the matrix.

Wireless World, May 1969



Fig. 6. The Venn diagram.



Fig. 7. The Venn diagram for A + B. The shaded area represents the required function.



Fig. 8. The Venn diagram for $A \ B \ \overline{C}$. The shaded area represents the required function.

bining A, B and C. Fig. 7 shows the Venn diagram for A + B, Fig. 8 shows the Venn diagrams for ABC.

Truth tables

The Truth table lists all possible combinations of the variables employed and takes the form shown in Fig. 9. Each variable has one column. The last column is reserved for showing the truth, or otherwise, of the function being illustrated for a particular combination of the variables concerned are true. This Truth table therefore represents the AND function.

Karnaugh maps

The construction of a Karnaugh map is shown in Fig. 10. Each variable is allocated half the area of a square, the other half of the square represents the complement of that variable. The positions that the variables occupy are shown in Fig. 10(a), (b), (c) and (d). The composite map is made up by superimposing these four divided squares upon each other as shown in (e). Sixteen squares result, each representing one unique combination of the variables, and each square differs from its neighbour by the negation of one variable.

Suppose one wished to construct a Karnaugh map for the function:

$\overrightarrow{ABCD} + \overrightarrow{ABCD} + \overrightarrow{ABCD} + \overrightarrow{ABCD}$

The map is drawn and 1 is placed in every square that represents one of the terms in the expression and an 0 is placed in all vacant squares. For convenience an addressing system is placed at the sides of the map so that each square can be easily identified. The map for the chosen expression is shown in Fig. 11. By definition all terms that are in adjacent squares differ only in the negation of one variable and may therefore be combined. All the 1s in our example are adjacent so all the terms can be combined. This is done by only selecting variables that are common to each adjacent square:

 $A\overline{B}C\overline{D}$ $A\overline{B}CD$ $ABC\overline{D}$ ABCD = AC

 $\therefore \overline{ABCD} + \overline{ABCD} + \overline{ABCD} + \overline{ABCD} = \overline{AC}$

A point which must not be overlooked is that squares on the right-hand side of the map are adjacent to squares on the left-hand side and the squares at the bottom are adjacent to squares at the top.

We will now return to the description of the instrument.

Obtaining the video signal

The three circles of the Venn diagrams can be drawn on the matrix-raster as shown in Fig. 12. To derive a signal corresponding to A it is only necessary to have a system of AND gates to select all the dots that are within circle A. And, in a similar fashion, signals corresponding to circle B and circle C can be obtained. These signals, A, B and C, are fed to front panel terminals for connection to external logic circuits. A video amplifier provides the Z drive for the oscilloscope and the input terminal of this amplifier is taken out to a front panel terminal which is labelled Z.

It follows if the terminal A is connected directly to the Z input then all the dots within the area defined by circle



D 0 0 1 1 С 0 1 1. 0 BA 0 0 0 0 0 0 1 0 1 0 1 0 0 0 1 1 1 1 10 0 0 0 0

Fig. 11. (right) A Karnaugh map

of a particular function des-

cribed in the text.

Fig. 9. Truth table for AND function.

Fig. 10. (left) The position of the variables in, and the construction of, a Karnaugh map.

200



Fig. 12. How a Venn diagram can be formed from the dots of the matrix-raster.

A will be visible on the screen. The same reasoning holds for B and C.

Now if the terminals A, B and C are connected to the inputs of an AND gate and the output of the AND gate is connected to the Z terminal, only the area common to A and B and C will be displayed. In other words the Venn diagram for the AND function will be shown on the screen.

The variables are negated to form \overline{A} , \overline{B} and \overline{C} so that the complement of the variables can also be made available on the front panel. Any logic gate or any logic circuit may be connected to the terminals and the function it performs will be displayed on the screen in terms of its Venn diagram.

In the Karnaugh map mode the situation is slightly more complicated because the information is presented on the screen as a pattern of 0s and 1s. The fact that four variables can be accommodated in this mode means, as mentioned earlier, that 16 characters have to be displayed at the same time as there are 16 possible combinations of four variables. In order that the displayed characters are separated, the output of the digital-to-analogue converters producing the scan waveforms are modified to produce the pattern shown in Fig. 13. The 16×16 matrix

•															
•	•		•	•	•										
•	٠		٠	•			•	•	•	•	٠	•	•	•	
•	•	•	•							•		•			
		٠													
٠	٠	•			•		•							•	
	•	•	٠		•	٠	٠	•	٠	•	•		•		•
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•	•				•										
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					•										

Tig. 13. How the matrix-raster is modified in the Karnaugh map and Truth table modes of operation to achieve character separaion.

)f dots is divided into 16 smaller 4×4 matrices each consisting of 16 dots. Each small matrix forms an area where a 1 or an 0 may be displayed.

The Karnaugh map, like the Venn diagram, consists of square in which various areas represent the variables; his was shown in Fig. 10. All the dots falling in an area 'epresenting a variable are selected by a series of AND gates, in the same way as was done for the Venn diagrams, so that four signals corresponding to the variables A, B, C and D are extracted. These are routed to the front panel terminals via the mode selection circuits.

These variables are connected to the logic gate or circuit that is to be demonstrated and the output of that gate or circuit is connected to the Z input terminal of the instrument. It follows that, as a Karnaugh map consists of every possible combination of the variables used, each of these combinations will be presented to the input of the logic circuit being demonstrated in turn as the spot scans the matrix. And, furthermore, when a particular combination is being presented the spot on the c.r.t. face will be scanning the area of the Karnaugh map representing that combination.

When the output of the logic circuit demonstrated is "up" it means that the section of the map that is being scanned by the spot is true for that particular logic circuit. It is arranged that an "up" signal to the Z input causes the character 1 to be formed in dots within the 4×4 matrix representing the combination of variables to the demonstration logic. When the Z input is "down", meaning that the combination of variables existing at that time is false, for the logic circuit being demonstrated, an 0 is formed in the area of the map being scanned.

The fact that a Truth table has various areas allotted to the variables is not so obvious as was the case for the Venn diagram and Karnaugh map. A Truth table for three



Fig. 14. How the variables are obtained in the Truth table mode.

variables consists of eight rows and four columns; an example was seen in Fig. 9. To show the complete table, $4 \times 8 = 32$ characters would have to be displayed at the same time. Unfortunately, as we have already seen, the maximum capacity of the instrument is 16 characters. To overcome this problem the table is displayed in two sections, one section when C is true and the other when C is not true. When only two variables are being used the problem does not arise because only 12 characters are needed.

The first two or three columns, depending on how many variables are being used, never alter as can be seen in Fig. 14 — the same pattern of 0s and 1s being displayed continuously. In Fig. 1 the logic to produce these columns is contained in the box labelled "truth table fixed format logic" which controls the 1 and 0 character generating circuits directly. The column 'C' of the truth table is controlled from a switch mounted on the front panel so that all 1s or all 0s, representing C or \overline{C} , can be selected.

The areas representing the variables are confined to the last column and are as shown in Fig. 14. The dots within these areas are gated out, as was done before, and used to form the output variables A, B and C. Whether a 1 or an 0 is displayed in the last, result column of the truth table is dictated by the logic circuit being demonstrated in exactly the same way as was done for the Karnaugh map.

A point which could lead to some confusion will now be cleared up.

Positive and negative logic

The Ferranti integrated circuits used are described as NOR gates in the manufacturer's catalogue. In this article they are referred to as NAND gates. The reason for this is that Ferranti used the negative logic convention when they specified the function that their circuits would perform and here, (as with previous) *Wireless World* articles we use the positive logic convention. Clearly the difference between the two conventions must be understood by the constructor who wishes to make the display aid. A brief explanation follows.

The difference between positive and negative logic can be summed up in one sentence. In positive logic the higher of two voltages represents 1 and in negative logic the lower of two voltages represents 1. The effect on the hardware of logic circuits of a change between the two conventions if profound.

The explanation here will start with the Truth table below. This was made-up by applying various inputs to an unspecified logic gate in an attempt to find out what sort of gate it was.

_	A	В	X
	0 V	0V	ov
	0V	+4.5V	ov
+4.	5V	0V	ov
+4.	5V	+4.5V	+4.51

Because we have not decided which logic convention to use, the inputs and outputs have been specified as voltage levels. As usual all possible combinations of the two input variables, A and B, have been covered. Now if we are working in positive logic—the higher voltage representing 1—we can reconstruct the Truth table by writing 1 for $\pm 4.5V$ and 0 for 0V. This is done below.

A	B	X
0	0	0
0	1	0
1	0	0
1	1	1

Examining this table we see that the gate we are testing gives an output only when A = 1 and B = 1, therefore X = AB, in other words the output is only true when both A AND B are true. Our once unspecified logic gate can clearly be seen to be an AND gate.

What happens if we work in negative logic? That is, the lower of two voltages is equal to 1. We have to reconstruct the truth table, writing 1 for 0V and 0 for +4.5V.

A	B	X
1	1	1
1	0	1
0	1	1
0	0	0

Examining this table we see that whenever A or B are 1 then the output is also 1, therefore X = A+B. This is the Truth table for an OR gate.

From the above it can be deduced that a gate that performs the AND function when the positive logic convention is used performs the OR function when the negative logic convention is used. When changing from positive to negative logic change AND to OR.

Selecting another gate, we will again produce three truth tables from it—that obtained by measuring voltage levels and those obtained by using the positive and the negative voltage convention. This has been done below.

A B X A B X A B 0V 0V +4.5V 0 0 1 1 1 0V +4.5V +4.5V 0 1 1 1 0	voltage		pos log	sitive ic	e	neg log	ativo ic	е	
OV OV +4.5V 0 0 1 1 1 OV +4.5V +4.5V 0 1 1 1 0	A	в	x	A	B	x	A	B	x
	0V 0V -	0V +4.5V	+4.5V +4.5V	0	01	1	1	1 0	000

In the positive logic case:

$$\overline{\mathbf{X}} = \mathbf{AB}$$

 $\therefore \overline{\mathbf{X}} = \overline{\mathbf{AB}}$ (negate both sides)

and $\mathbf{X} = \overline{\mathbf{AB}}$ (double negatives cancel)

In positive logic the gate produces the NAND function.

For negative logic:

$$\mathbf{X} = \overline{\mathbf{A}} \overline{\mathbf{B}}$$

 $X = \overline{A + B}$ (De Morgan's Theorem)

So in negative logic the gate performs the NOR function.

Next month: the digital-to-analogue converters will be described. The general method of construction and the integrated circuits will be discussed.

Since this article was written development work on the display aid has continued and a number of additions have been made that greatly extend the instrument's usefulness. However, it is stressed that the extra facilities result from additions, rather than modifications, to the basic circuits. The reader is therefore advised to build the basic instrument first and then decide which of the options he wishes to add.

Many versions of the instrument are possible, ranging from the basic instrument; through a version with only one control switch (on/off) that will produce simultaneously the Venn diagram, Karnaugh map and Truth table on the same oscilloscope for any external logic circuit; to an instrument bristling with 32 push-button switches that will produce four separate displays on the same instrument. Each of the display areas can be switched individually to show a Truth table, Karnaugh map or Venn diagram. Also, with this instrument, two external logic circuits can be accommodated at the same time and any of the four display areas can be individually switched to show either external circuit number 1 or number 2 or the difference between the two circuits. This latter facility is useful for demonstrating De Morgan's theorem and for logic card testing. Finally, all four areas can be switched to the Karnaugh map mode to form a single Karnaugh map of six variables.

Calculation Time Saver

This units converter "slide-rule" has been designed by Wireless World specially to suit the needs of the electronics or radio man. It will be available at an advantageous price exclusively to readers of this journa —details next month.



From the microelectronic point of view the theme of this year's Salon des Composants Electroniques was undoubtedly a move into the domestic consumer market by most of the main integrated circuit manufacturers. Television seemed to be the area of major interest with audio amplifiers coming a close second, and even the motor car and the camera did not escape the attention of some enterprising manufacturers.

Major reasons for this apparent volte-face are that only recently have the set makers accepted that microelectronics can offer them any real economic advantages and, having catered well for the industrial market, the .c. manufacturers are looking elsewhere in order to sell on the broadest front possible.

It would now appear, for instance, that a complete stereo tuner-amplifier including he i.f. strip, stereo multiplex decoder, audio pre-amplifiers and main amplifiers can now be built from monolithic integrated circuits. Such an equipment would probably fall into he medium quality class offering up to 5W per channel, or if one were to have hybrid hick film main amplifiers, up to 15W per thannel.

It will probably be only a matter of time before the i.c. manufacturers attack the ery high quality audio market and we will be hearing the arguments for and against nonolithic sound (or mono-stereo?) as igainst transistor and valve sound. And takng things to their ultimate conclusion, as he manufacturers of m.o.s. l.s.i. circuits ppear to be on the look-out for new applicaions, perhaps in a few years we shall see the programmable hi-fi system. This might conain a number of tape cassettes with digitally oded information on the tape identifying ach piece of music. A programming system vould allow any piece of music to be selectd and played back in any order-the possiilities could be almost endless.

Many more medium scale integration m.s.i.) m.o.s. integrated circuits were to be een in manufacturers' catalogues, and it is pparent that much work is being done to ncrease the speed of this type of circuit. Ithough offering many advantages in the adustrial process control field and in any ther application where speed is not paranount, such as desk calculators, m.o.s./ n.s.i. arrays will not replace e.c.l. circuits in omputers for a long time to come, However, is possible that m.o.s. circuitry will start to eplace large ferrite core stores in a couple of ears as it is expected that m.o.s. dynamic storage will become cheaper than its ferrite counterpart after taking into account the cost of the drive circuitry.

Discussing this application, a representative of Texas Instruments said that it is possible that four chips, each with a one to two thousand bit storage capacity, could be mounted in the same package, and if a suitable package was designed and the beam lead interconnection technique was employed, it would be possible to replace an individual chip within the package should it fail.

Also on the industrial front some examples of m.o.s. complementary arrays were seen.

The U.S.S.R. was represented at the Salon for the first time this year and a wide range of r.t.l., d.t.l., t.t.l., m.o.s., monolithic and hybrid integrated circuits for industrial and domestic applications was on show, although no examples of l.s.i. were seen. A representative said they were exhibiting in Europe to gauge the general reaction to their products and that they may exhibit in the U.K. sometime this year. Does this herald the entry into the world microelectronic market of another large contender?

The first example of a product employing l.s.i. intended for what could be considered the consumer market was shown by Schneider Radio and Television and was presented as their contribution to what they call "price decrease technology". The product, a low-cost digital multimeter, will be marketed in the U.K. by Honeywell Ltd. (Hemel Hempstead) in the near future.

The instrument, known as the Digitest 500, employs a single chip which incorporates a three-stage decade counter, all the



Schneider digital multimeter

decoding logic for the display, all of the instrument's control logic and part of analogue-to-digital converter. The chip is manufactured by General Instruments Europe and it is certainly an achievement to incorporate all these functions within the limitations imposed by a 16-lead dual-in-line flat pack.

The instrument has 17 ranges and is capable of measuring a.c. and d.c. voltage and current, and ohms. The price of the meter is expected to be about $\pounds 110$ in this country.

Digressing slightly for a moment, it is worth while mentioning a novelty item seen on the Schneider stand that could be called an innovation awaiting an application. A Swiss engineer, Herr Vogel, has designed a multitrack automatic tape replay mechanism intended for inclusion in a digital voltmeter, or other instrument operating in the b.c.d. code, that gives an audible, as well as a visual, indication of the quantity being measured. On the Schneider stand the tape mechanism was seen built into a digital voltmeter and the combination was called "Voltmétre Numérique Parlant".

Microcircuits for television

The main area of activity as far as integrated circuits for television is concerned was in f.m. i.f. amplifiers. SGS have produced one (type TAA661) which is suitable for operation at 6 or 10.7MHz so that it may be employed in television or radio receivers.

In this circuit after three stages of amplification and limiting the signal is split into two. One signal, now a square wave because of the limiting, is fed directly to a discriminator which is a coincidence detecting circuit. The second signal is fed to an external tuned circuit, the output of which is a sine wave. This sine wave is also fed to the coincidence detecting discriminator.

As the applied modulation varies the frequency of both the sine wave and the square wave vary in unison; however, the sine wave will be subjected to a phase shift which depends on frequency because of the tuned circuit. The net result is that the output of the coincidence detector will be a series of pulses the mean value of which are proportional to the modulation. The discriminator circuit, together with explanatory waveforms, is shown in Fig. 1.

The TAA661 will operate with a supply

voltage from 4.5 to 15V, 12V at 15mA being typical. It has a frequency range 5kHz to 60MHz and will provide a 60-dB gain at 5.5MHz; a.m. rejection is typically 40dB with a modulation frequency of \pm 50kHz to a depth of 30%.

Another f.m. i.c. (type TAA710) was shown by Intermetall; this incorporates an oscillator, mixer, i.f. amplifier and discriminator and requires very few external components. It has an a.m. rejection of 40dB and requires an input between 1mV and 1V.

The French subsidiary of the Philips group, La Radiotechnique-Compelec (R.T.C.) announced a frequency modulated i.f. amplifier which included a discriminator and a variable gain i.f. stage. The a.f. output voltage could be varied by altering the potential at one of the i.c. input pins.

R.T.C. is currently developing an integrated decoding matrix (TAA 470) for colourtelevision receivers which produces the R, G, and B signals from Y, R-Y, G-Y and B-Y.

Integrated voltage regulators intended for supplying tuning potentiometers in varicaptuned receivers were shown by both R.T.C. (TAA550) and Intermetall (ZF33, ZTK33). In all cases the output voltage was 33V.

A colour decoder (MC1325) in a dualin-line flat pack announced by Motorola produces the R-Y, G-Y and B-Y signals from the composite chroma signals and the two reference phases.

Hybrid microcircuits are usually designed for a particular customer application and are not generally sold as standard production items. There are, of course, a number of exceptions to this, voltage regulating circuits being an example. N.S.F. Telefunken are producing three thick film hybrids for television use. These are a PAL flip-flop, an a.f. input and driver amplifier which is intended for use with an external output transistor to provide up to 1W, and an a.f. amplifier with a 50-mW output intended to drive an external complementary output pair. This latter amplifier has an open circuit voltage gain of > 5,000 and a distortion factor of 1%.

Plessey have seven i.cs for television at present under development, these include a colour decoder, i.f. amplifier and a synch separator.

.... for audio

The most powerful microcircuit audio amplifier seen was a thick film hybrid made by Bendix, that could deliver 15W r.m.s. into a 3Ω load at about 1% distortion at 1kHz. The microcircuit is housed in a ceramic package (5 × 2.5 × 0.7 cm) that is designed to be attached to a heat sink. The amplifier has a 60-dB power gain and requires a 350-mV input for full output.

A monolithic amplifier in a dual-in-line package with special arrangements for heat sinking was announced by General Electric (PA246). This amplifier provides 5W r.m.s. into a 16 Ω load. The distortion performance depends on the external components employed; however, a typical figure is 0.7% although this can rise to as high as 5% with careless selection of the external components. The 3-dB points of the frequency response are 30Hz and 100kHz at 2.5W output.

A large number of manufacturers were showing audio amplifiers with 1 or 2W outputs and it is not proposed to mention all of them as some of these have already been described in *Wireless World*.

Motorola manufacture two 1-W monlithic audio amplifiers offering different performances. The better of these (MC1554G) is designed to operate with a 16 Ω load and is capable of offering only 0.4% total distortion at full output power with suitable selection of external components. Under these conditions the frequency response is flat from about 50Hz to 500kHz. Because of the very wide bandwidth very great care must be taken to keep all wiring as short as possible and to avoid stray coupling between input and output to prevent v.h.f. instability. RCA introduced an interesting microcircuit (type CA3048) which houses four independent a.c. amplifiers in a dual-inline flatpack. Each amplifier has a minimum 53-dB gain, an open loop bandwidth of 330kHz, an input impedance of 90k Ω and will provide 2V output at low distortion.

Fig. 2 shows the CA3048 connected as a complete stereo preamplifier, the gain is 46dB and the total harmonic distortion at 2V r.m.s. output at 1kHz is < 0.2%. Another audio application for this i.c. would be as a mixer.

Secosem-Recherche described work they had been doing in applying the piezo-m.o.s. effect to gramophone pickup arms. An experimental model produces 50mV output but later it is hoped to increase this by a factor of ten with improved mechanical coupling.

.... for radio

A major, but unfortunately publicity-shy, British manufacturer who will have to remain anonymous, mentioned at the exhibition that they are producing a complete car radio as a single chip. This in itself is not outstanding as other firms announced fully integrated radio chips. The difference is that the firm in question has provided its car radio with a 4-W audio amplifier on the same chip and seems to have solved the problem of local heat generation and the difficulty of heat spreading through the chip and upsetting earlier circuits.

A complete monolithic stereo multiplex decoder in a dual-in-line flatpack (MC1304) was shown by Motorola. A choice of a plastic package (suffix P) or ceramic package (suffix L) is available. This circuit requires three external coils and a few assorted resistors and capacitors. A 200mV r.m.s. multiplex input signal is required and the left and right channel audio information is available at the outputs. An output is also provided for a 12V, 40mA, stereo indicator lamp. The

Int



Fig. 1 A monolithic discriminator

average channel separation is 34dB and the cotal harmonic distortion is typically 0.5% with a 1% maximum.

R.T.C. showed a monolithic i.c. (TAD100) intended for use in a.m. portable radios; it contains the mixer, local oscillator, i.f. ampliier and a.f. preamplifier and is designed for operation from a 9-V supply.

Most of the i.f. amplifiers for television sound described in the television section could be used in f.m. radio receivers. Plessey showed a range, the 600 series, of nicrocircuits intended for use in comnunication receivers.

.... for the many

dicrocircuits for unusual applications were o be seen on several stands. For instance ntermetall had a monolithic i.c. for motor ar flashing direction indicators. This was rranged to operate the direction lamps in he usual way; however, should one of the imps fail the frequency of flashing of the iternal indicator lamp doubled to provide a varning. A switch allows all lamps to be lit at he same time for signalling purposes.

Also for the motor car industry General istruments of Europe are developing a rake control system that will assist in reventing skidding on wet or icy surfaces. his will be an m.o.s. /l.s.i. device that counts ilses from generators driven by the four ad wheels. The number of pulses received ver a given period is compared with the ontents of a store which holds a number of ilses proportional to the velocity of the shicle over the same period. Outputs reilting from the comparison are used to introl individual wheel brakes to achieve aximum braking efficiency without skidng. Early tests of the system have been ost encouraging.

Intermetall showed a miniature 1.1-V volge stabilizer intended exclusively for the ock and watch industry.

R.T.C. have an integrated circuit AA500) designed to provide the necessary

270

100 u

impedance transformation when telephone handset carbon microphones are replaced with crystal microphones. The circuit has automatic volume compression and has two input leads and two output leads and does not require a separate power supply. A later version of this circuit will shortly be introduced which incorporates a diode bridge so that the polarity of the connections is of no importance.

Also from R.T.C. a monolithic circuit (TAA560) intended for automatic cameras. A light cell provides the input to the circuit which then automatically adjusts the cam-'era's shutter speed.

Both R.T.C. and General Electric had m.o.s. divider circuits primarily intended for electronic organ applications although many other uses are of course possible.

.... for industry

A very high-speed read and write "scratch pad" memory unit organized as 16 words each of one bit was shown by RCA. The memory, type TA5318, which has a typical readout time of only 7ns, employs e.c.l. circuits and is housed in a dual-in-line flatpack. The operating temperature range is -55° to $+75^{\circ}$ C.

RCA also announced a new addition to their range of co.s./m.o.s. (complementary symmetry metal oxide semiconductor) elements which is described as a dual complementary pair plus inverter (CD 4007). This element has a fan out of 50, low 'l' and '0' output impedances and a propagation delay of about 35ns. It may be connected as a triple inverter element, a three input NOR gate, a three input NAND gate, a high sink current driver, a high source current driver, a high sink and source current driver or as a dual bi-directional transmission gate; additionally it will perform the relay tree-logic function.

R.T.C. had a quadruple p-channel m.o.s. element (TAA530) designed for use as a chopper. Offset voltage is quoted as

Vcc

 $2\mu V(max)$ with a maximum drift of 20nV/°C and offset current is 2nA(max) with a maximum drift of 20pA/°C.

The designer wishing to use m.o.s. /l.s.i. arrays has three main courses of action open to him. He can select circuits from the wide range available as off-the-shelf items in manufacturers' catalogues, or he can have a special-purpose element made by specifying the interconnection pattern to be used on standard array of m.o.s. devices, finally, he can have a special-purpose l.s.i. chip designed and manufactured to suit his application.

Coincident with the exhibition Marconi-Elliott released details of work they have been doing using a computer to satisfy the designer's third choice mentioned above. Basically an engineer works with a light-pen and graphic display coupled to a Marconi Myriad computer. The computer store holds details of all the standard m.o.s. circuit elements, additionally the engineer can use his lightpen and display to design special-purpose circuit elements. Again using the light-pen and display the engineer can lay out all the circuit elements on the chip and specify the interconnection pattern. The information in the computer store is then used to control a plotter which draws the masks and a cutand-strip machine which cuts the masks needed to produce the device.

A large range of read-only memories is available in m.o.s. technology. The one with the highest capacity seen was from Fairchild; this could store 4096 bits (type 3502). This device has 7,000 transistors on a single chip and represents a storage density approaching 700,000 bits per square inch. Also from Fairchild was a seven segment character generator for c.r.t. displays. It is well known that the numerals from 0 to 9 can be constructed from seven lines or segments. The Fairchild device (3250) generates all the combinations of the seven lines necessary to produce the numerals and the c.r.t. scan voltages as well.

In conclusion it is perhaps worth while mentioning, for those who do not already know, that the m.o.s. device was invented and patented some 15 years before the transistor by J. E. Lilienfeld and O. Heil. The lack of sufficient knowledge of semiconductor physics at the time, followed by development of valve, and later transistor technology, resulted in the m.o.s. lying dormant.



Three-dimensional Television

Holography offers possibilities for analysis and reconstruction of stereoscopic images in monochrome and colour

by R. Brown

A successful three-dimensional television system has long been a dream of many people in the television field; but such a system would be very difficult to construct and until recently none of the techniques available has appeared to meet with wide acceptance among the viewing public. Recently, however, the technique of holography has been arousing great interest as a means of producing threedimensional photographs without using special spectacles, and it seems worth considering whether a holographic television system could be produced. If it could, the received picture would be a tremendous advance on existing two-dimensional systems. In addition to being three-dimensional without the use of special glasses it would have several other characteristics not found in existing 3D systems. If a viewer moved to the right or left objects in the foreground of the picture would move relative to more distant objects so that it would be possible to look around them to see what was behind. In addition, a viewer would have to refocus his eyes to look at distant objects after looking at close objects. In fact, the picture would be



Fig. 1. Making a hologram. The point object to be recorded reflects some of the laser light back towards the photographic plate in expanding spherical shells. There it interferes with the plane waves coming from the laser via the mirror. indistinguishable from the original scene. The scenes produced from existing holograms do indeed bear an uncanny resemblance to the original scene, and, what is more, they can be produced in full colour as well as in black and white.

Making a hologram

To understand how a hologram is made it might help if we first took a look at what happens when we view a scene. Any object, however complex, can be thought of as being made up of many thousands of small points. Each of these points, when the object is illuminated by the sun or by artificial light, reflects light in all directions. Each point can indeed be thought of as a point of source from which light waves radiate in expanding spherical shells.

The ideal recording system would be one in which the viewer was presented with exactly the same information in exactly the same form as when looking at the original scene. In other words he should be presented with replicas of the expanding spherical shells of light waves coming from that scene. Ideally, we should dispense with a lens and find some means of "freezing" the light waves from the scene in a photographic emulsion and then "unfreezing" these waves at a later time so that they can continue on towards the eye of the observer. If this can be done the observer will "see" the scene just as it would appear to him had he looked at it directly. This is just what we can do with holography.

Photographic emulsions can certainly record the amplitude of the light waves; but they cannot record that other essential characteristic of a wave--its phase. They do, in fact, respond only to the amplitude or intensity of light waves; but to any engineer it must be clear that one way in which the phase information can be recorded is by using interference effects. This is, however, impossible with conventional light sources because even the so-called monochromatic light sources are in fact generators of a quite wide band of light wavelengths which have random phase relationships. We can, however, produce suitable interference effects by using a laser, which gives a highly coherent monochromatic beam.

To record light waves directly without a lens, the scene or object is illuminated by laser light and some of the light from the laser is also directed straight on to a photographic plate where it interferes with the light reflected from the scene.

If we take as an example a single point object it can be seen from Fig. 1 that the spherical wavefronts travelling from the point interfere with the plane wavefronts coming direct from the laser. At points where the two wavefronts are in phase the light intensity in the photographic emulsion will be a maximum: at points where the two wavefronts are 180° out of phase they will cancel out and there will be a minimum. Thus there will be a series of dark and light areas on the photographic emulsion. It can be seen from Fig. 2, that in the case of a single point object the spacing of these light and dark areas increases from the top to the bottom of the plate. The brightness of the light areas depends upon the intensity of the light reflected from the point and the spacing of the light and dark areas depends upon the distance of the point from the plate. Thus the recording contains all the information about both the brightness and the distance of the point object. If the object consists of a number of points the light waves reflected from these points will set up a number of different sets of interference patterns and, what is more, the spacing of the lines in the interference patterns will depend upon the relative distances of the points.

To reconstruct an image of the point object the photographic plate is first developed so that the light areas become clear and the dark areas opaque. Then it is illuminated with a laser beam coming from the direction of the original reference beam used during the recording of the hologram. The light passesthrough the clear areas and each clear areacan be thought of as a thin line source oflight. The light spreads out from the manyline sources and interference occurs betweer the many different sets of light waves. It car be seen from Figs. 3 and 4 that there will be at least three different sets of wave fronts on the far side of the photographic plate as a resul of this interference!

An attenuated version of the illuminating beam will continue on in the same direction because the light from the different opening is in phase in that direction. There are also two other beams going off at angles to this beam One of these other beams goes off in a direc tion in which the wavefronts of the lighfrom one opening is in phase with the previou

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wavefronts of the light from the adjacent opening. The other beam is in a direction in which the light from one opening is in phase with the light in the preceding wavefront coming from the adjacent opening.

There is an important difference between these latter two beams. It can be seen from Fig. 5 that both wavefronts are spherical but in the case of the top beam the wavefronts are expanding and in the case of the lower beam the wavefronts are converging. In fact, the wavefronts from the top beam are exactly the same as those that were reflected from the point object during the recording processes. If an observer positions himself on the far side of the photographic plate, these waves will enter his eye and he will "see" the original point object on the far side of the plate. What is more, the position of this point object relative to the plate will be the same as the position of the original object relative to the plate during the recording processes: the observer will indeed see the point as though he was looking through a window in the position of the photographic plate, or hologram as it is called. If the object recorded were more complex, the observer would see each of the points of which it was composed in their proper positions. He would, in other words, see a three-dimensional image of the original object.

The converging spherical wavefronts in the lower beam eventually concentrate into a spot. A photographic plate placed there will record an image of the point object without the need for a lens. Again, if the object is a complex one all the different points in it will be reconstructed and a photographic plate will record an image of the object.

Full colour holograms

Full colour holograms can be made fairly easily if rather expensively by making use of the fact that the angle at which the reconstructed images can be seen depends upon the angle of the reference beam used in the recording process. We can, for example, make a full colour hologram by directing red, blue and green laser beams on to the photographic plate at different angles. The same laser beams also illuminate the scene, of course. To display the full colour image recorded, the hologram is illuminated by red, blue and green laser beams coming from the same angles as during the recording process. If this is done the red, blue and green images are reconstructed at the same spot and the observer sees a full colour image. Each laser beam will also reconstruct images from the interference patterns of the other two beams but these images will be in a different place and if the image is viewed through a suitable mask these spurious images will be blocked.

A more interesting and economically more attractive way of producing full colour images nvolves directing the three reference beams on to the back of the photographic plate during the recording process.¹

Holographic television

What, then, do we need to do to transmit tolograms by television? First of all, to televise a scene it must be illuminated with high ntensity pulsed laser light. Probably several



Fig. 2. When the plane wave coming direct from the laser meets the spherical waves reflected from the point object, constructive and destructive interference occurs. This produces a series of bright and dark fringes whose spacing increases down the photographic plate.



Fig. 3. When a hologram is illuminated with a laser beam three new beams are formed. An observer in the position shown will see the original point object in the position shown. The photographic plate will record an image of the point object without the need for a lens. The third beam is an attenuated version of the illuminating beam.

phase locked lasers would have to be used because the picture would be very harsh if only one laser was employed. Pulsed operation is essential because the scene must remain almost perfectly still during the exposure time of each frame, otherwise the interference pattern will be smeared and made useless. It has been suggested by Professor Leith of Michigan that movement during the exposure must be kept down to less than one-quarter of the wavelengths of the light.² Movement at rates of up to about six metres per second could be allowed if the pulse length were restricted to about ten nanoseconds. This means that it would be possible to record and reproduce images of a person walking quite quickly, but movement much faster than this might present some problems. However, there have been very great strides in the development of high-power short-pulse lasers in recent years and speeds much greater than six metres per second, which was quoted at a meeting in Montreal several years ago, will soon be possible.

There is another rather serious limitation —the relatively short coherence length of lasers. Clearly, there must be no abrupt changes in the phase of the light from the laser between the time the light illuminating the front of the scene has left it and the time



Fig. 4. When the simplest hologram of all, one in which the light and dark fringes are equally spaced, is illuminated with a laser beam the interference between the light getting through the spaced openings produces the three beams in the manner shown in 1, 2 and 3. that the light illuminating the back of the scene has left it. In other words, the coherence length of the laser beam, that is the distance between points with a constant phase relationship, must be longer than the depth of the scene to be televised. Unfortunately, pulsed ruby lasers, for example, have a coherence length of only several inches and even the best continuously operating gas lasers have coherence lengths of only six feet or so. This means that the width of the scene that can be viewed is limited to a few inches if a ruby laser is used and to a few yards if a gas laser plus a high-speed shutter is used.

Again, however, steps are being taken to reduce the seriousness of this problem. The Radio Corporation of America's Princeton



Fig. 5. How the expanding and converging wavefronts are produced when the hologram of a point object is illuminated.



Fig. 6. Reducing the bandwidth requirements by discarding most of the information contained in the hologram. Each of the lenses in (a) selects a small part of the area of the hologram, (b), and magnifies it so that it fills the space allocated to that lens in the plane indicated. It is the very coarse fringe pattern in the plane that is scanned by the electron beam and transmitted. laboratories have recently demonstrated a gas laser in which one of the mirrors is replaced by a piezoelectrically driven interferometer. This greatly increases the coherence length of the laser and holograms of scenes of much greater depth can now be recorded.

A conventional television camera can be used to convert a black-and-white hologram into a video signal suitable for transmission. The thick holograms recorded with the reference beam on the far side of the recording surface might present more serious problems, but no doubt a solution to them can be found.

The most difficult problems in the way of a successful television transmission system arise. from the very large bandwidth requirements. E. N. Leith has calculated that to transmit all the information in a single colour hologram would require about 30,000 times as much bandwidth as a conventional television transmission system.³ This is clearly an impossible requirement at the present time and might still be impossible even if millimetre or laser wavelengths could be used for the transmission link. However, all the detail in the hologram does not necessarily have to be transmitted. If it were, the picture would have a much higher definition than present television pictures. We can thus greatly restrict the amount of information transmitted and still have a threedimensional picture every bit as detailed as present day two-dimensional pictures.

Several ways of doing this are being investigated. It is, for example, possible to make use of a very attractive feature of the hologramthe fact that the information from every small point in the scene is spread over the whole of the photographic plate. This means that we can tear the hologram in half and still extract the entire scene from either half. This is why computer manufacturers, for example, are very interested in the idea of holographic memories. Tears and scratches have little or no effect because all the information is contained in the undamaged parts of the film. This, of course, is quite unlike conventional photographic memories where even a tiny blemish can destroy important information.

One could, in principle, simply transmit only the information contained in a tiny central area of the hologram. However, the size cannot be reduced far enough because definition is lost as the size is reduced, and too much detail is lost before the bandwidth requirement comes down to a practical figure. In the United States the National Aeronautics and Space Administration has developed a technique in which a metal mask containing a large number of small holes is placed in front of the hologram at the sending end and a similar mask is placed in front of the reproducing system at the receiving end.⁴ In this technique tiny samples spread over the whole area of the hologram are transmitted. This is sufficient to reduce the bandwidth requirements to practical limits and yet leave a reasonably detailed picture.

An alternative way of dividing a hologram into a number of elements and then discarding all but a tiny portion of each element has been developed by the Bendix Corporation W. E. Kock, who joined Bendix from N.A.S.A. has pointed out that the optical interference patterns on a hologram correspond to a definition of 50,000 lines an inch, so that on a 4-inch square hologram there are about $(4 \times 10)^{10}$ picture elements as compared to the 250,000 elements present in a 500-line square television picture. He proposes having one small lens for each of the basic pictur elements.5 As shown in Fig. 6, these lense will image a few fringes at the centre of each of the basic areas of the hologram into greatly magnified form that will fill th elemental area in the plane indicated. Produc ing the tens or hundreds of thousands of lense may not be quite the formidable task it migh appear because Mr. Kock has devised a way o producing tiny zone plates that can perforr just as well by a photographic technique.

The coarse fringe pattern formed in th plane (Fig. 6) is scanned by an electroni beam as in conventional television and th resulting signal is transmitted to the receiving end. There the pattern in the plane *i* reconstructed and a second array of lense de-magnifies the fringes back to their tru optical fringe size. In this way a great deal c the information is discarded but again a 31picture of acceptable definition is obtained

Holographic TV receiver

So far as the receiver is concerned the sign circuits would be quite conventional up to point; certainly the extra electronic processir circuits would not present any serious pro lems. Also, once the hologram had been r produced it could be illuminated with a las or white light to reproduce the scene in the same way as standard holograms are illumi ated. The real problem is going to be repr



Fig. 7. A holographic television receiver. The intensity of the laser or electron beam (depending upon the system) is varied in accordance with the hologram interference pattern. This reproduces the hologram pattern on the screen which is then illuminated by the illuminating laser and the viewer sees reconstructed 3D images.

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ducing the interference pattern. J. Upatnieks, a colleague of E. N. Leith, told the Montreal meeting referred to above that there appeared to be two techniques which after considerable development might be used to reproduce the hologram.

One of these techniques is based on the Fischer system which was developed in the early 1940s in Geneva by Fritz Fischer as the basis of a large screen theatre television receiver.^{7 8} It consists basically of a thin film of oil between two systems of shutters which are placed between the light source and the screen. When the oil film is perfectly smooth the shutters completely block the light. To produce a picture, the film is scanned by an electron beam which distorts its surface. This diffracts light past the system of shutters and illuminates the screen in the appropriate places to produce the picture. In principle there would seem to be no reason why some modification of the basic Fischer system could not be used to produce the characteristic hologram interference pattern.

The other proposed solution to the problem uses photochromic glass. This glass has been known for over a century and possesses the useful characteristics of darkening when exposed to light or other electromagnetic radiation close to the visible part of the spectrum. In a hologram television receiver, this darkening could be produced by a laser beam scanning across the glass from side to side and up and down in much the same way as the electron beam scans a conventional cathode-ray tube screen. Varying the intensity of the laser beam would vary the amount of darkening of the glass and so the hologram patterns could be reproduced.

There are several hundred photochromic compounds known. They all involve the use of atoms or molecules which are bistable in that they can have two states with different atomic, molecular or electronic configurations. The molecules are colourless in their normal state and darken as the result of switching over to their other state when light shines upon them. When the light is removed they switch back to their original state.⁹¹⁰

Silver chloride glass sensitive to the ultraviolet region of the spectrum might be suitable for use in a holographic television receiver. It is thought that the colouring effect in this glass is due to the formation of neutral silver _atoms as in ordinary photographic film. All the radiation is then captured by the silver. This process is non-reversible in the case of photographic film; but in photochromic glasses, because of the extremely small size of the crystals, the process reverses when ight is removed. The much greater volume of he crystal in a photographic emulsion encourages the neutral silver atoms to aggregate nto stable colloidal particles; but this does 10t occur in glass because the crystals are about one sixty-millionth the size of the photographic emulsion crystals. Other factors, uch as the impermeability of the glass, are lso important. At the moment all of the photohromic compounds available are slow acting by the standards required for a television ystem. But, as yet, there has been no incentive o develop fast acting photochromic glass.

All the basic requirements for 3D colour elevision would seem to be present; but there s equally no doubt that the problems that remain to be solved are very formidable. Even if holographic television was technically feasible it is very unlikely that we would see it in use for broadcasting for several decades because of the high cost of the equipment. Nevertheless, there are many applications where the transmission of even still 3D pictures would be very useful. The advantages of a having a 3D television system on a space probe on the surface of, say, Mars, which could transmit still pictures only, would, of course, be very great. This perhaps explains the N.A.S.A. interest in the technique.

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Announcements

The International Aerospace Instrumentation Symposium erroneously announced for March 23–26 at the College of Aeronautics, Cranfield, will in fact be held on those dates next year. Further information may be obtained from N. O. Matthews, Department of Flight, The College of Aeronautics, Cranfield, Beds.

The Manchester Electronic Instruments Exhibition organized for the past two years on behalf of a group of 16 manufacturers is being enlarged this year but will again be held at the Hotel Piccadilly, from September 23rd to 25th. The organizers, Industrial Exhibitions Ltd, are this year arranging a similar exhibition (restricted to 16 companies) at the Hotel Leofric, Coventry, from September 16th to 18th.

A new M.Sc. degree course in microwave and communications engineering, organized jointly by the Electrical and Electronic Engineering Departments of the Universities of Leeds and Sheffield, will begin in October 1969.

An international conference on **earth station technology** for satellite communication, sponsored by the I.E.E., I.E.R.E. and other institutions, is to be held during October 1970 at the I.E.E., Savoy Place, London W.C.2. A new standards laboratory installed by G. & E. Bradley Ltd., at its Neasden, London, factory, has been approved by the British Calibration Service for a wide range of measurements at radio frequencies. This is the first laboratory to obtain B.C.S. approval for r.f. calibration.

The Radio Advisory Service, which was established jointly by the London Chamber of Shipping and the Liverpool Steam Ship Owners' Association some years ago, is no longer a separate organization. It recently became the Radio and Navigation Department of the Chamber of Shipping, 30/32 St. Mary Axe, London E.C.3.

The names of two RCA companies in the U.K. have been changed. RCA Ltd, of 50 Curzon Street, London W1Y 8EU, will in future operate as RCA International Ltd. The tide RCA Ltd has been adopted by what was RCA Great Britain, of Sunbury on Thames.

Corning Glass International, S.A., the Belgian-based subsidiary of Corning Glass Works of America, have opened a London Office at 3 Cork Street, W.1. Corning subsidiaries with facilities in England include Electrosil Ltd in Sunderland and Miniature Electronic Components Ltd in Woking.

Motorola Inc. is to establish a further subsidiary company in Britain with the title Motorola Automotive Products Ltd. The company will be located at Stotfold, near Hitchin, Herts, and will provide complete engineering, production, marketing, sales and service facilities for the Motorola eight-track stereo tape player for cars.

Brookdeal Electronics Ltd, 2 Myron Place, Lewisham, London S.E.13, manufacturers of signal recovery equipment, have appointed J. Arndt Jensen, Kongevejen, Allerod, Denmark, and Into Oy, P.O. Box 10153, Helsinki 10, Finland, as agents for their products.

B & K Instruments Ltd, 59 Union Street, London S.E.1, have been appointed distributors of the range of variable frequency filters, variable frequency a.c. power sources and laboratory power amplifiers manufactured by the Krohn-Hite Corporation, of America.

G. A. Stanley Palmer Ltd, Island Farm Avenue, West Molesey Trading Estate, Surrey, have been appointed agents for Arco, of Bologna, Italy, manufacturers of electronic components.

The digital systems department of Ferranti Ltd, has been awarded a contract to supply **SHAPE Technical Centre** at The Hague, with an FM1600B computer and associated peripheral equipment. The computer will be used in experimental and research work.

A contract for the design, construction and flight testing of a full-scale experimental **airborne early warning radar** has been placed by the Ministry of Technology with Elliott-Automation Radar Systems Ltd.

International Marine Radio Company Ltd, of Croydon, have received an order worth over £120,000 for marine communication and navigation equipment from Scottish Ship Management Ltd. The equipment will be installed in 12 new bulk carriers on order from British and Norwegian yards.

The Marconi Company has won an order valued at almost £250,000 to provide remote-controlled, highfrequency radio communications equipment at Nandi Airport in Fiji.

The London office of Marconi International Marine Co. Ltd, is now at 30/34 New Bridge Street, E.C.4. (Tel: 01-236 8113; Telex 884729).

The marketing department of Racal Instruments has moved from Crowthorne, Berkshire, to Bennet Road, Reading, Berks. (Tel: Reading 85571.)

V. N. Barrett & Co. Ltd., suppliers of high-vacuum and scientific equipment, have moved to new premises at 1 Mayo Road, Croydon, Surrey, CRO 2QP.

Fluke International Corporation have moved to larger factory and office premises at Garnett Close, Watford, WD2 4TT.

Cadmium Nickel Batteries Ltd has moved from factories in Park Royal to premises at Castle Works, Station Road, Hampton, Middx. (Tel: 01-979 7755).

News of the Month

Skynet terminal delivered

military satellite com-Skynet, the munications system, will begin to operate early in 1970. Two satellites, one in operation and the other a standby, will move in a synchronous orbit 23,000 miles above the Indian Ocean. Of the nine earth stations to be employed, five will be fixed, two (to be built by Plessey) installed in the assault ships H.M.S. Fearless and Intrepid, and two will be capable of transportation by air. The whole system will provide interference free communication for British armed forces, and nation-wide communication from the Atlantic to the Far East, including Hong Kong.

The satellites and launchers, and some specialized control and monitoring equipment, are being built in the U.S.A. under an agreement with the U.S. Government which allows the U.K. to benefit greatly from the American space investment. The satellites contain some British designed equipment and are capable of meeting the conflicting requirements for communicating to large and small earth-stations simultaneously. The operating functions of the satellites, and their positions in the sky, will be controlled

7 metre diameter Skynet dish

from the U.K. command and monitoring station at Oakhanger, Hants. For long-life (3-5 years) switchable duplicate equipment has been installed in each satellite. The first satellite will be launched late this summer.

G.E.C.-A.E.I. (Electronics) has handed the first of the four air-transportable earth stations over to the Ministry of Defence. The terminal was designed and built in only 18 months.

Each terminal is made up of three basic sections each light enough to be carried by standard transport aircraft and helicopters. Once 'on-site', six semi-skilled men can erect it, and 'lock' it on to the satellite's wavelength, in only three hours. Each station's 7m diameter reflecting dish provides a number of voice channels, and is assembled from 12 petal sections.

Base-band and i.f. circuitry is housed in a control unit separate from the aerial structure. From the unit, in the transmit mode, processed information at the i.f. enters a travelling-wave tube amplifier feeding a klystron. The standard power output is about 5kW. For the reception of satellite signals a Ferranti liquid-nitrogen-cooled pa-

* "Computer-'designed' Circuitry, Wireless Wor July 1966 p.373.

rametric amplifier is employed.

The three other stations will be completed by the middle of this summer.

At Christchurch, Hants., another terminal will support the Skynet project by making highly accurate measurements for initial calibration and testing of the spacecraft in orbit.

Computer-aided design presses on

A massive four-day conference on computeraided design-the collected papers weigh 1b-has just ended at Southampton University. A short while earlier Racal Research Ltd announced that the electronic c.a.d. service called "REDAC" which they started in 1966* now has 50 major British companies using it. At the time of launching there was some scepticism about the value of such a service in electronics design work and, indeed, individuals in Racal feel that it might have been rather "oversold", but since then the Ministry of Technology has come in as a partner, a good deal of experience has been gained and the service has apparently consolidated itself.

The Racal establishment, at Tewkesbury, Glos., consists basically of a digital computer (Elliott 4130), with means for transmitting and receiving customers' data (Datel 200, Telex, phone, post) and 25 engineers for translating customers' electronic design requirements into forms suitable for computer processing. The engineers are also continuously engaged in up-dating information stored in the computer (mainly componen parameters); developing computer prog rammes; making component measurements producing equivalent-circuit devic "models" suitable for computer operations training customers; and other tasks to de with the day-to-day development of the ser vice. At present there are 31 computer programmes available to customers in a manua Some of these programmes are directly con cerned with particular types of electroni circuits. For example, one of them calculates, and if required optimises, the corr ponent values of an active RC filter to give required bandpass response with gain. Th programme gives a full nominal component list and the results of worst-case d.c. and a. analyses. Other programmes are of mor general applicability, such as calculating th harmonic content of a waveform over specified band of frequencies, or computin the elements of a hybrid- π model of a trai sistor from the measured 'y' parameters.

The main justification for operating suc a service is, of course, economic—the savir of engineers' time on the innumerable calculations that should be done (but often aren' to ensure that a reliable design is obtaine This is particularly important in tolerancin sensitivity analysing and optimising the cor ponent values of electronic circuits intende for mass production. To do such wo properly one must calculate a complete s of performance figures (e.g. gain, frequence

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response, noise) for every possible value within tolerance of every component or device in the circuit—a huge task far beyond the capabilities (or willingness) of a computerunaided engineer. The main limitation of the service seems to be that the customer concerned with active circuits is at present restricted to a small range of particular circuit topographies—those for which computer programmes have been written—and if he has some other configuration outside of this range his work cannot be handled immediately, or at least fully.

In the June issue we hope to report on some aspects of the Southampton conference referred to above.

E.I.D. apprentices

The four-year apprenticeship scheme introduced by the Electrical Inspection Directorate (Ministry of Technology) in 1955 with five boys now has an annual intake of over 70. The total number of trainees going through the centre is about 250. Recently the title of E.I.D. was changed to Electrical Quality Assessment Directorate (E.Q.D.) but the programme of basic training at the school which forms part of the E.Q.D. headquarters at Aquila, Bromley, Kent, remains unchanged.

The apprentices are normally placed in courses on block release at Bromley College of Technology and follow the National Certificate course (electronics or telecommunications) or City & Guilds technician course. In addition to a full electronics training the -apprentices also cover basic mechanical engineering practices.

The necessary qualifications for entry to the apprenticeship scheme is that the boys nust be studying for G.C.E. 'O' level or C.S.E. in mathematics, physics, or an acceptable science subject, and must be between extern and eighteen years of age. The period of apprenticeship may be extended in respect of those boys who wish to complete Higher vational Diploma, Bachelor of Science or other advanced academic courses. At present ive are taking B.Sc. and four H.N.D.

At the annual prize-giving on April 2nd he recipient of the Rowland Memorial Cup for the best electronic work of the year was A. J. Smith, of Maidenhead, Berks.

Computer speeds aircraft blind landing experiments

A new development in the blind landing exerimental programme being carried out by the Blind Landing Experimental Unit at LA.E., Bedford, involves the use of a mall-scale analogue computer. It is incororated into a standard autopilot in a Comet B aircraft and in conjunction with a variety f sensors and an inertial platform, it enables apid changes of equipment parameters to be nade during investigation of a wide range of ontrol laws. The computer, model TR48 upplied by Electronic Associates Ltd, is ruggedized" to withstand the rigours of



Interior of the Comet 3B showing the E.A.L. computer and patchboard

airborne use. "Automatic Landing in Airline Service" was the subject of a 6-page article by R. E. Young, in the November 1967 issue of *Wireless World*.

The current automatic landing situation is that Tridents of B.E.A. and Super VC10s of B.O.A.C. have been making automatic landings on scheduled services in Europe and North America although always in conditions of good visibility. The ultimate objective is to make the blind landing so safe and reliable that aircraft can operate in any weather conditions, even dense fog, with no reduction in movement rate, i.e. no change in the flight schedule. Experiments are being conducted in an area within the restricted straight-in approach at a shallow angle of 3°, dictated by international agreements, using the existing instrument landing system. Those now in progress are concerned with improving existing control laws and at present a system is being studied which mixes inertial terms with the main i.l.s. guidance signals. The i.l.s. is susceptible to interference. especially that caused by reflections when other aircraft are taking-off in the path of an aircraft making a landing, and inertial mixing provides the possibility of making the flight control system less susceptible to such interference.

The difficulty at R.A.E., until use was made of the TR48 computer, was that in order to introduce a new or different factor into a test flight, the autopilot electronics required to be dismantled for wiring modifications, to provide a change of operating characteristics. Now, with the computer, the required parameters can be programmed before take-off and switched-in during flight. A second TR48 computer is installed in a flight simulator in a ground laboratory and programmed via a large patchboard. When a successful landing has been made on the simulator with the required inputs fed into the autopilot system, the actual patchboard is then taken from the simulator and connected to the computer in the aircraft for a

comparative flight test. Recordings of flight activities for subsequent analysis are made on a 14-track tape recorder and on a chart recorder.

The computer is used in an experimental role only and one would not be installed in operational aircraft. At some stage in development of blind landing, safety standards acceptable to airline operators and the Air Registration Board could be optimized in the autopilot which would then be constructed in conventional size and housed in the equipment bay in the normal way. In all, well over 20,000 blind landings have been accomplished by the B.L.E.U. at Bedford and at London Airport. The A.R.B.'s safety target is that there should be not more than one fatal accident in 10 million landings. Although control performance is the concern of R.A.E., equipment reliability would be the responsibility of the manufacturers. It was said during a demonstration of the blind landing equipment at Bedford that if aircraft movement could be reliably maintained in all weathers it would be possible to dispense with the excess fuel carried to cater for a possible diversion from the destination airport. As an example of the possible saving involved figures were quoted from the Concorde's estimated performance. The passenger payload of this aircraft is only 6% of the all-up weight but the excess fuel carried is 10%.

Home-made X-rays

According to a report in the Daily Telegraph a two-year survey of 5,000 colour TV receivers in Long Island, U.S.A., has shown that 20% of them are emitting potentially dangerous X-rays. Of 37 different makes of receiver at least one of each make was found to be radiating at more than 0.5 milliroentgens an hour which is said to be the danger level established by the National Council of Radiation and Measurement in 1960. The report did not state if the radiation was emanating from an area on the receiver to which viewers are normally exposed.

Optical fibre telecommunications

The Post Office, having staked its claim to an early, God-given medium of communication ("Let there be light"), is now thinking seriously of using it for the largebandwidth telecommunications systems of the future. At present the optical fibre waveguide seems to be the most attractive way of conveying the light waves, according to a paper given by F. F. Roberts (P.O. Research Department) at an I.E.R.E. conference on lasers and opto-electronics at Southampton University. In common with the piped optical and millimetre-wave systems, it offers a transmission attenuation substantially independent of bandwidth "up to any bandwidth of interest". All three systems are expected to be cheaper per channel-kilometre than coaxial cable for bandwidths greater than certain break-even values. But the optical fibre system should be "appreciably easier" to install than the other two in a congested country like Britain. The main problem, discussed at length by Roberts, seems to be in finding a suitable optical fibre material that will keep the transmission losses (absorption and scattering) down to a practical figure-provisionally set as a target of 20dB per kilometre of waveguide.

What is envisaged at present is a waveguide consisting of a 5μ m-diameter oxide glass central core surrounded by a 100μ m-diameter cladding of similar material but of different rafractive index. Optical waves of $600-1,000\mu$ m (the red end of the visible spectrum) would be launched into this guide by a solid-state laser (e.g. gallium arsenide), propagated in the HE₁₁ mode, and received by a solid-state diode (e.g. silicon). At intervals along the transmission path there would be repeaters, each containing a diode receiver, a solid-state amplifier and a laser transmitter. The bandwidth available would be several GHz for distances of about 1km between repeaters.

Waveguide of this kind should not be confused with the fibre-optic "light-pipes" used in punched-card readers and other viewing applications: these are very much thicker and have thousands of different modes of wave propagation; consequently, because of interference effects, their bandwidth is less than 1MHz for a 1km length of material.

Amateur cloud-cover pictures

In 1967 Ivor le Mercier (4S7LM), president of the Radio Society of Ceylon, set about designing and building a receiver to enable him to record cloud-cover pictures transmitted by the satellite Nimbus-2.

The receiver line-up was as follows: two r.f. stages (AF139), mixer (OC171), crystal oscillator (OC170), tripler (OC171), four 10.7MHz i.f. stages (OC171), two a.f. stages (OC71) and an output stage (OC810). The receiver functioned well and required $0.8\mu V$ for 27dB quieting.

The aerial consisted of a 6-turn helix supported as a wooden boom 3.5m (11.5ft) long and a 1.9m square (6ft) ground plane of galvanized mesh. The aerial was mounted in gimbals, firstly being roughly aligned with the satellite's orbit and then swung by hand to track the satellite. A monitor loudspeaker

Paul Voigt's contributions to audio were the subject of a meeting of the British Kinematograph, Sound and Television Society at the Royal Institution of Great Britain on March 26th. His work in sound recording was discussed and demonstrated and here Ralph West is shown demonstrating an early Tractrix horn from which Voigt developed his famous corner horn. On the left is Peter Walker who chaired the meeting. Paul Voigt, who is 67 and now lives in Canada, sent a recorded message.





One of the pictures recently received by Ivor le Mercier

mounted near the aerial facilitated this process.

Synchronization was achieved using a phase locked oscillator at 2400Hz, divided down to 20Hz, based on a circuit by W. G. Andersen which appeared in the November 1965 issue of QST.

Unfortunately by the time the equipment was completed Nimbus-2 was out of commission; however, in June 1968 picture: from ESSA-6 were received and recordec satisfactorily.

To produce complete cloud-cover pictures, signals from the receiver' discriminator are recorded on tape and "played back" through an oscilloscope fitter with a 35mm camera.

Student paper contest

Kenneth Gray, a research assistant at Wool wich Polytechnic, has been chosen to rep resent the United Kingdom and Republic o Ireland Section of the I.E.E.E. at the Institu tion's regional student paper contest in Mor treux, Switzerland, on 23rd May.

The Regional winner will later compet with other Regional winners in New York t determine the overall winner.

Mr. Gray's paper describes a Fourier me thod of investigating transient acoustica spectra and its application to human speech and is relevant to the problem of speec communication with computers.

New recording process ?

A member of Wireless World's staff recentl received some literature from Reader's Dige: inviting him to buy a set of gramophor records ("Mood Music for Listening an Relaxation") made by RCA engineers usir an amazing new process called "Cyclophon Sound". The RCA people in Britain wer unable to discover anything about it. Fallir back on etymology, one finds that "cyclo from the Greek, means circular, or perhap in this case, rotating; while "phonic", al from the Greek, means relating to soun The word "Sound", one must assum means sound-though perhaps a differe kind of sound, not quite the same as the "phonic" sound. From this analysis one dduces that the new process involves som thing that is circular, perhaps rotating, at produces sounds. What could it possibly be?

Operational Amplifiers

4. Applications

by G. B. Clayton,* B.Sc., A.Inst.P.

In instrumentation a need sometimes arises for a precise d.c. voltage reference or for a variable direct voltage that can be set accurately with a calibrated potentiometer. In such cases it is necessary to avoid loading the voltage standard if its stability is to be ensured. Also, a potentiometer must not be loaded if its dial calibration is to be accurate. The high input impedance and low output impedance of an op. amp. used as a voltage follower makes it valuable as a buffer in such



cases. In the circuit shown R_z is used to set the current through the reference zener at that value for which the temperature coefficient of the zener is a minimum. R_2 and R_1 are adjusted to compensate for olerance in the zener voltage and so obtain he precise voltage required at the output. The RC filter may be added to attenuate noise and pick up.

Calibrated Potentiometer Voltage.



The high input impedance of the follower bove does not load the potentiometer and perefore does not affect its calibration.

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Current Sources

A requirement that sometimes arises is that a load be supplied with a constant direct current that is linearly related to some input voltage. A single transistor used in the manner shown is often employed to supply a relatively constant current.

Single Transistor Current Source.



In applications requiring a higher degree of precision than is possible with the single transistor circuit, op. amps. may be used to advantage as current output devices. In cases where the output current capability of the op. amp. is insufficient a booster amplifier may be added.

 $i_L \simeq E_B/R_E$

Current Source. Follower Connection (Floating Load).



Features. The circuit is similar to the follower with gain with the load replacing the feedback impedance. The load must not be earthed. All follower connections are of course subject to common-mode limitations; in this case the input voltage e represents a common-mode signal and it must not be allowed to exceed the rated maximum for the amplifier.

Current Source, Inverting Connection (Floating Load).



With the error voltage $e_{\varepsilon} \rightarrow 0$ the voltage across R2 and R3 must be the same

$$IR_2 = I_3R_3$$
 and $I = \frac{c}{R_1}$

The load current

$$I_{L} = I + I_{3}$$
$$I_{L} = I \left(1 + \frac{R_{2}}{R_{3}} \right) = \frac{e}{R_{1}} \left(1 + \frac{R_{2}}{R_{3}} \right)$$

Features. The circuit uses a form of the inverting amplifier and draws an input current $I = e/R_1$. The load is above earth as before; it may be a complex load. The circuit can be used as a deflection coil driver.

Current Source (Earthed Load).



Features. In addition to the negative feedback applied to the inverting terminal of the amplifier, the circuit uses a positive feedback loop to the non-inverting terminal to achieve a very high effective output impedance and thus a constant load current. One side of the load is conveniently at earth potential. Two input voltages may be used, both referred to earth. If only one input signal is used and it has the appropriate polarity, it is preferable to earth the eg terminal and use the e_1 terminal as the current determining input. The input current drawn from the source e_1 is determined by the resistor R_1 which can be made quite large to limit the input current. If e_2 is made the current determining input the shortcircuit load current is drawn directly from e_2 through R_2 .

Current to Voltage Transducer.



Features. The circuit is essentially the inverting amplifier configuration with the input resistor omitted. It represents a simple but convenient method of current measurement. With the error voltage $e_{\varepsilon} \rightarrow 0$ the circuit introduces negligible input voltage drop but the output voltage is developed at the low-impedance, high energy capability output terminal of the amplifier. The high current sensitivity of the circuit makes it useful as a null detector. Null detectors are often used off null, and under these circumstances this simple circuit would saturate and give little indication as to how far off null it was. When used as a null detector it is convenient to arrange a modified non-linear response characteristic (see right).



Features. The inverting amplifier is used with the feedback resistor replaced by a device exhibiting a logarithmic characteristic. Log. amplifiers require particular attention to bias current and input offset adjustment and to choice of logarithmic device if reliable operation is to be achieved over several decades. Interchanging the position of the logarithmic device and the input resistor, as shown below, gives an antilog response.

Antilog Amplifier.



Combinations of log and antilog amplifiers may be used to generate a variety of functions. The principle of operation of a log multiplier is illustrated above. Log Multiplier.







 R_0 is a non-linear resistor which decreases for increase in current through it. Sensitivity is maximum near I = 0

Null Detector (Log. response).



$$e_{d} = e_{0} \frac{R_{1}}{R_{1} + R_{2}}$$

Hence $e_{0} = \left(1 + \frac{R_{2}}{R_{1}}\right) e_{d}$
 $e_{0} = \text{const.} \left(1 + \frac{R_{2}}{R_{1}}\right) \log_{\cdot 10} \frac{|1|}{I_{0}}$

Features. The circuit makes use of the approximate logarithmic relationship that exists between current and voltage for a diode-connected transistor.

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Corrections

"Simple Class A Amplifier"

Three small errors crept into the artic published last month. In Fig. 8, showing the modified circuit for high input impedance the rail-dropping resistor should be 3.5 (2.2k for stereo) not 39k (22k for stereo). the last line of Table 3 the transistor corplement should be 2 \times MJ480 and 2 2N697 to correspond with Fig. 9 (b). Thphotograph on page 152 showing layout of single channel has been wrongly lettered wirespect to Tr_1 connections. Reading up frethe bottom the correct letter order is *e*, *b*,

Although in Fig. 3 the output transisto are boldly marked MJ480, the $15 - \Omega$ ve sion requires MJ481s with a 2N1613 driver. This was stated, perhaps a little o scurely, at the end of the third paragraph (page 150.

"Acoustic Absorption Materials"

A printer's error reversed the meaning of t sentence at the end of the first paragraph page 173 (April). It should read "It is n easy to ensure"

"Why not angular frequency?"

Mr. Whitehead miscalculated the freque cies for 5,000 and 10,000 rads/sec in 1 letter on p. 178 (April); they are 796 a 1592 Hz.

Is there a future for the

Television V.H.F. Bands?

A survey of the British television scene as it affects viewers

It has been widely reported in the radio trade press that a large proportion of the viewing public was unaware that when BBC-1 and I.T.A. programmes become duplicated on u.h.f, these programmes will be transmitted in colour. It would be instructive to learn how many people were aware that the programmes were to be duplicated anyway and what advantages they thought it had for them. For someone who is receiving a perfectly satisfactory 405-line picture on v.h.f. (and the broadcasting authorities' figures say that 98 to 99% of the population do) to be told that he will require a new receiver and aerial within the next seven years to receive the same programmes on u.h.f., he may feel that his viewing arrangements are being interfered with unnecessarily and moreover will involve him in additional expense. (The introduction of BBC-2 was different-it was a new programme.) Unless the viewer buys a colour receiver, the picture quality will be only -marginally better than before. But to tell the public that the change is being made solely to enable colour programmes to be -received would not be quite true, although it is true that without making the change to 625-line u.h.f., colour prog--rammes of any sort cannot be seen. This is because of a -Government decision and not because it is technically impossible. It will be recalled that the Independent Television Authority was at one stage leading protagonist for a 405-line colour -service and gave some impressive demonstrations in an attempt to prove that the 405-line colour picture quality was of a high order.

If the viewer cannot afford a colour receiver, he can still receive BBC-1 and I.T.A. programmes on 405-lines v.h.f. -undisturbed for the next seven years, because it is intended to -keep the existing transmitters operational for that period following the start of duplication, and if he has already -purchased a dual-standard receiver for the purpose of watching BBC-2, the technical features of the changeover should, in -theory, be of no consequence to him. He should already be using a u.h.f. channel group aerial the bandwidth of which is designed to cover the new BBC-1 and I.T.A. u.h.f. channels for -his area as well as the existing BBC-2 channel, so that it should -merely be a case of tuning to the new BBC-1 and I.T.A. -frequencies on the u.h.f. scale.

This may not be quite as straightforward as it sounds. Although all u.h.f. transmitters for any particular area will be co-sited, and of the same order of radiated power (the most far-sighted arrangement made in recent years), the propagation conditions may not be identical on each channel. Whereas the multi-channel u.h.f. aerial was initially aligned and adjusted for optimum reception of BBC-2, it may require re-adjustment for compromise reception of three channels, involving the -services of an aerial rigger. In any case, viewers have had no previous standard by which to judge their reception of BBC-2, but it may be a different story when they make the switch from -v.h.f. BBC-1 and I.T.A. to u.h.f. BBC-1 and I.T.A. It should be remembered that while duplication of programmes on v.h.f. and u.h.f. lasts, people with dual-standard receivers will at any time be able to make a direct comparison between them. Those whose BBC-1 and I.T.A. reception is not so good on u.h.f. will feel justifiably aggrieved when v.h.f. 405-line transmissions are eventually switched off. Then again, at some date in the future, after the three programmes are established from his main transmitter for the area, the viewer may find that his particular locality has been provided with a relay transmitter to improve reception (perhaps making it comparable to his v.h.f. reception). This will be on another channel and necessitate a new aerial which will have to be mounted in a different plane to the one used for reception from the main transmitter.

It has been argued that long before the end of the seven-year duplication period the v.h.f.-405-line-only viewer will require a new receiver anyhow, so that it will simply be a matter of his purchasing a replacement receiver which will operate on the new 625-line u.h.f. standard. A complete change from v.h.f. to u.h.f. may still mean a new aerial will be required at a cost of, say, $\pounds 12$.

The whole or part of this cost could be offset by the saving on the single-standard receiver with its simpler circuitry than the dual-standard type. The changeover switch and other components associated with dual-standard operation will be eliminated. The dual-standard receiver has always been regarded as a compromise design not giving of its best on either system, whereas the single-standard version will have a frequency response tailored to give optimum performance on 625 lines only and it will be inherently more reliable. There will still be a need for a v.h.f. /u.h.f. tuner unit in some areas perhaps for reasons discussed later, but in the main a u.h.f.-only tuner will be all that is necessary.

Despite this the receiver industry seems to be wary of offering a single-standard u.h.f.-only receiver in advance for those areas where all three existing programmes will be available on u.h.f. and on 625 lines because at the present time no British black-and-white receiver of this type is on sale, and one major manufacturing group is advising its dealers not to stock single-standard receivers this year. This same company predicts that it will still be manufacturing dual-standard receivers in seven years' time. A spokesman for a second major set-making group told W.W. that it will be a bold P.M.G. who declares a large number of domestic receivers obsolete by announcing the end of v.h.f. TV broadcasting and his company foresees the continued use of v.h.f. even in ten years' time. It is to be hoped that the single-standard receiver will become a reality in time.

In view of what has been said it may be worthwhile re-examining the question of why the change to 625 lines and at the same time the change to u.h.f. became necessary, and where it is likely to take us in the future.

In 1956, Parliament asked the Television Advisory Com-

mittee (T.A.C.) to say whether the existing 405-line standard was likely to remain adequate for the next 25 years, and whether there was any reason why the 625-line standard, broadcast in u.h.f. Bands IV and V, should not be used for broadcasting in the U.K. if it were recommended by the C.C.I.R. as the European standard. Following large-scale field trials the T.A.C. recommended that 625-line broadcasts on u.h.f. should be adopted if Europe generally adopted this. At a European VHF/UHF Broadcasting Conference in Stockholm in 1961, most countries concerned decided to adopt as standard an 8MHz channel bandwidth which greatly eased the problem of channel sharing with neighbouring countries. It is to some extent because of the variety of channels in the v.h.f. bands in Europe and the varied spacing of sound and vision carriers within them that interference from long-distance stations in these bands is troublesome. The presence of this interference which is due to "Sporadic E" effects is also the reason why 405-line v.h.f. colour transmissions are not considered practicable. Because of the introduction of a colour subcarrier, the visible effects of this type of interference would increase for about nine million viewers during active periods. Should the uniformity of 8MHz channelling agreed at Stockholm be disturbed, a serious interference problem could also arise with colour transmissions, in Bands IV and V, as would be the case in v.h.f.

The T.A.C.'s findings were published as a report in 1960¹. The report stated that 405-line standards would not be adequate for all purposes for the next 25 years and that a 625-line standard would give a definite improvement in picture quality particularly with larger screens (our italics). While nobody will dispute that an increase in the number of scanning lines per frame will reduce the visibility of the line structure, the subjective improvement in picture quality should be relatively better irrespective of screen size. Viewing distance plays a part here. Returning to the 1960 report, this pointed out also that the introduction of television broadcasting into Bands IV and V would present the last opportunity the U.K. would have of changing its line standard. In other words, it was a case of now or never. If 625 lines were used in Bands IV and V, then this system would eventually have to be used in Bands I and III. But what of Bands I and III, will they ultimately become available for other programmes? This is a question which remains unanswered.

To advise on the future of broadcasting, the Government set up a committee in 1960 under the chairmanship of Sir Harry Pilkington (The Pilkington Committee) which considered the 1960 T.A.C. Report and weighed the non-technical factors involved in changing the line standard. After deliberating for about two years, this committee finally presented its report² to the P.M.G. on June 5th, 1962. It recommended that the standard be changed to 625 lines and found that the increased costs of transmitting and receiving on the new standard both to the viewer and the broadcaster, were not significant. The committee also recommended the "duplication" method of changeover which virtually puts Bands I and III out of use for some seven years as far as new programmes are concerned. It is interesting to note that had the Pilkington Report recommended the retention of 405-line transmissions, the number of channels which could be used in Bands IV and V would be no greater than on 625 lines in view of the U.K.'s previous undertaking to adopt 8MHz channel spacing. And the number of national programmes possible is related to the number of available channels

The T.A.C. Report recognized that to produce a u.h.f. replica of the present v.h.f. services is virtually impossible and eventually some of the v.h.f. channels will have to be employed to extend the four u.h.f. programmes to those areas where u.h.f. reception is poor. In order to reach anything like the population covered by Bands I and III with transmissions in Bands IV and V, some 60 main u.h.f. transmitters are planned,

supported by hundreds of relay stations. No one knows at this stage the exact number of relay stations required (the I.T.A says 400), but unless the countryside is littered with low-power stations, amounting in some places to almost a transmitter to each village, some viewers are bound to find that their u.h.f reception is inferior to that obtained on v.h.f. The feasibility o using so many stations is only possible by the development o the unmanned station technique, otherwise the number o technical manning staff required would be unreasonable. Poo reception on u.h.f. may not be confined only to remote area either.

According to estimates the final coverage on u.h.f. will b 95% of the population as compared with 98-99% coverage or v.h.f. This may seem to fall not far short of identical coverage but 1% represents $\frac{1}{2}$ million people. It is at this stage, one is to assume, that use will be made of some of the then redundan v.h.f. channels to complete the coverage of 625-line prog rammes. This operation together with the institution of two new programmes on 625-line v.h.f. is referred to vaguely by th Pilkington Report as "re-exploitation of Bands I and III" although it is not known what this will entail, pending decision from the Government.

When considering the question of the future use to whic. the v.h.f. bands should be put it has to be kept in mind that th Pilkington Report recommended that ultimately six televisio programmes should be planned for: four on u.h.f. and two a v.h.f. Since the 5MHz spacing of the present 405-line channe' allows just sufficient number of channels (13) to give nationcoverage with two programmes, it will be physically impossibility to provide the same number of channels with the agreed 8MH channel spacing unless Band III is widened. If Band III were t be extended from 216 to 222 MHz then Bands I and II= together would provide 9 channels which could provide tw programmes, although the coverage attainable would be les than that of the existing services on 405 lines. Gaps could t filled in by making use of u.h.f. Whatever use is eventuallmade of Bands I and III, after the withdrawal of BBC-1 an I.T.A. programmes we shall be left in a few years' time with network of v.h.f. transmitters which could straight away giv national coverage with one additional programme, operating o 9 channels with 8MHz spacing. It is unthinkable that th broadcasting authorities would relinquish their claims fc television broadcast facilities in the v.h.f. bands, and remembe ing that the B.B.C's and I.T.A's charters expire in 1976 plar for "re-exploitation of Bands I and III" should emerge lon before that date.

In the meantime pressure is being exerted by all radio-use for more frequency space in an already overcrowded spectrum and the v.h.f. Bands I and III would, for example, partiallfulfil the needs of mobile radio operators if their plea for mor space at u.h.f. is rejected. Then again the Conservative part has announced recently that if returned to Parliament they wi set up a a further 100 radio stations which for part of the tin at least, according to the announcement, will be broadcasting a v.h.f., straining further the resources of frequency space. Th Government will need to seek advice from the P.M.G's Fr quency Advisory Committee and although in this counti-Band V has not yet been wholly allocated and further secto could be made available for television it would be a pity if the present v.h.f. bands were lost to television broadcasting. In ar event, in accepting the Pilkington Report, Parliament has al accepted the implications of demands on frequency spa required by six television programmes and if this is to be th future pattern, it is now up to Parliament to provide the necessary operating frequencies and authorize the capit expenditure.

J.H.V

References

¹Report of the Television Advisory Committee 1967. H.M.S.O. 1968. ²Report of the Committee on Broadcasting 1960. H.M.S.O. Cmnd 1753.

Digital Microcircuits

A description of the major logic families

by D. E. O'N Waddington,* M.I.E.R.E.

Microcircuits have been with us for a few years now and the trend in new equipment s to make more and more use of them. This s not entirely because they are small but because they offer very definite advantages n cost and reliability as well as simplifying lesign procedures. In order to distinguish between the different families of circuits, the nanufacturers have found it convenient to lescribe them by the initial letters of their rircuit descriptions. Each family has its own particular characteristics, advantages ind disadvantages, and before selecting a amily for a particular application it is as -vell to appreciate the differences. For this eason a glossary of terms and abbreviations s included at the end of the article. This is by no means comprehensive as only the well nown logic families are included.

Resistor-transistor logic (r.t.l.)

This was one of the first logic forms to be uilt into integrated circuits; probably recause it is simple and was already in use vith discrete components. The basic circuit f a two-input gate is shown in Fig. 1. When either, or both of the inputs is taken ositive, the output will go to 0 V-either r_1 or Tr_2 , or both, will be bottomed. Iowever, when both of the inputs are held t o V, the output will be positive-both r_1 and Tr_2 will be cut off. Typically these ircuits operate from a low voltage, 3 V eing about normal; however, under certain ircumstances it is permissible to increase he supply voltage. Two types of r.t.l. are ade, a low power type-2 mW/gate with propagation delay of 30 ns and a higher ower type-12 mW/gate with a propaation delay of 12 ns. Noise can be a problem ith r.t.l. as, with a logic swing of 0.8 V, he noise margin is only 0.3 V.

iode-transistor logic (d.t.l.)

his type of logic, together with t.t.l., omes under the general heading of "current nking logic". In practice this means that the switching action at the input to a gate onsists of transferring the current holding the gate "on" into the circuit driving it. In ther words, sinking the current in the utput stage of the driving circuit. Most of the current sinking logic forms are com-

Typical integrated circuit characteristics

logic family	gate delay ns	fan out	mW per gate	toggle rate MHz	noise margin V	power supply V
r.t.l.	12-40	4-5	2-12	3-25	0.3	+3
d.t.l.	8-25	8-25	9-30 (1)	5-25	1	+4 - +5
t.t.l.	6- <u>12</u>	6-15	10-25 (1)	12-30	1	+5
e.c.l.	2+6 (*)	25	40-110	20-120	0.35	-5
m.o.s.	50-300	_	5-20	1-5	4-8	24

(1) 1mW types are available but are very much slower in operation.
 (2) gate delays of 0.9ns are now available.









Fig. 3. Basic i.c. d.t.l. gate.

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patible and can be interconnected. However, the fan-in and fan-out conditions must be checked as the loads for the different circuit types may be different.

The circuit of a d.t.l. gate in integrated form amounts to an almost direct transfer of the discrete component circuit shown in Fig. 2. The circuit operates as follows:— When the inputs are held positive the diodes D_1 and D_2 are reverse biased so that current flows through R_1 , D_3 and D_4 into the base of Tr_1 keeping Tr_1 bottomed. When either input is connected to earth, the current through R_1 will be diverted through either D_1 or D_2 to earth and Tr_1 will switch-off.

The switch-on noise margin is determined by the voltage required at "A" for base current to flow in Tr_1 . As the voltage drop across a forward biased sllicon diode is approximately 600 mV, the voltage must be at least:

$$V_{D_3} + V_{D_4} + V_{BE} (Tr_1)$$

= 3 × 600 mV = 1.8 V.

This means that the voltage at either input point must be at least one diode forward voltage drop less than that at point "A" if the gate is to change state. Thus this noise margin will be approximately 1.2 V. The switch off noise margin is determined by the power supply and is usually about 2 V.

There is one major disadvantage of translating this circuit into integrated form. The turn-off time depends largely upon how fast charge can be pulled out of the base region of Tr_1 . In discrete component circuits this is achieved either by making D_3 and D_4 slow recovery types so that they provide a transient low impedance path from the base of Tr_1 to earth, or by returning R_2 to a negative supply rail. Neither solution is very good for integrated circuits; the

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former because it is not easy to make slow and fast diodes in the same chip and the latter because it is nicer to have only one supply rail. Another approach is to make the value of R_2 relatively low. The disadvantage of this is that it would rob base current from Tr_1 thus reducing fan out. The integrated circuit solution is shown in Fig. 3. D_3 is replaced by an emitter follower, Tr₂, which gives the necessary current gain so that the value of R_2 can be reduced without affecting the fan out. Sometimes times the point "A" is brought out as a connection to the integrated circuit so that additional diodes can be connected to expand the circuits fan-in capability. The main limitation is one of speed, as each diode adds capacitance across the input, thus increasing the switching time.

It is worth noting that it is often possible to improve the turn-off time of the circuit by using the spare fan-out. For instance, if a circuit has a fan-out of eight and is only driving two loads, there is a six-load capability which can be utilized. Thus, if one load is equivalent to $4 \ k\Omega$, the output collector load can safely be shunted by $4/6 \ k\Omega$. This will improve the turn-off time at the expense of additional "on" current.

Transistor-transistor logic (t.t.l.)

This logic series is primarily an integrated circuit form which has no counterpart in discrete circuitry. The two features, which are peculiar to t.t.l., are the multi-emitter input transistor and the "totem pole" output. The latter circuit, although recognized as a design feature of t.t.l., has been used both in r.t.l. and d.t.l. circuits.

The basic gate is shown in Fig. 4. If all the emitters of Tr_1 are connected to the positive rail, the base-emitter diodes are reverse biased while the base-collector diode is forward biased thus supplying base current to Tr_2 switching it on and, at the same time, switching Tr_4 on and Tr_3 off. The switching-off of T_{r_3} is assured by D_1 (See Fig. 5). When any one (or more) of the input emitters is connected to earth, the corresponding base-emitter diode will conduct and Tr_1 behaves like an ordinary transistor and "bottoms", switching Tr2 off, Tr_3 on and Tr_4 off. Because the turn-off time of saturated transistors is longer than the turn-on time, there will be a very short period, usually of only a few nanoseconds duration, when Tr_3 and Tr_4 are both conducting. The circuit will, therefore, draw a pulse of current limited only by R. from the supply. For this reason adequate supply decoupling must be used. The great advantage of this form of output circuit, over that normally used in d.t.l., is that the turn-on and turn-off times are roughly equal. With d.t.l., the turn-off is generally longer than the turn-on and rapidly becomes longer if the load on the output is capacitive. However, the resistive output has two advantages over the "totem pole" which should not be forgotten. Firstly it is possible to directly connect together the outputs of two or more gates to give a logical "OR" function. "Wired" or "dot OR".) The other advantage is that the

actual output potential, when the output transistor is switched off, is equal to the supply line voltage. With the "totem pole" circuit, however, the output voltage will be at least 0.6 V less than the supply voltage. Some t.t.l. circuits are now being manufactured without the "totem-pole" output so that the wired "OR" function can be achieved.

Emitter-coupled logic (e.c.l.)

Like r.t.l. and d.t.l., e.c.l. was also used in discrete form long before it was incorporated into integrated circuits. However, because transistor matching is possible in integrated circuits, this logic form is very well suited to integration. Unlike all the circuits previously described, e.c.l. is a nonsaturating logic form, that is, the logic functions are performed by current steering rather than by saturated switches and, therefore, the delays associated with turn-on, turn-off and hole storage no longer apply. Instead the "switching" time is limited by the speed of the transistors, making e.c.l. one of the fastest forms of integrated circuit logic available.

The basic gate is shown in Fig. 6. Unlike the previous circuits, in which the gates were always of the inverting type, e.c.l. usually provides both output polarities, i.e. output-1 is inverting while output-2 is non-inverting.

If either, or both, of the inputs is taken positive, Tr_1 or Tr_2 , or both will conduct, cutting Tr_3 off. That is, the current through R_2 will be steered through R_1 . This will cause output-1 to go negative and output-2 to go positive. If both of the inputs are taken negative, Tr_1 and Tr_2 are switched-off and the current through R_2 is steered through R_3 . This will cause output-1 to go positive and output-2 to go negative. Saturation is avoided by a proper choice of the ratios of resistors and reference voltage. In practice the logic swing is only about 800 mV with a noise margin of about 200 mV. Despite this low noise margin, e.c.l. is a practical system as the noise generated by the operation of this type of gate is very much less than that generated by any of the saturating logic forms. The main advantage of e.c.l. is that it, and e.²c.l., are the fastest logic form available. Propagation delays of less than 2 ns are obtainable and some manufacturers even predict delays of less than 0.5 ns in the future.

Metal-oxide-silicon logic (m.o.s.)

This is very different from any of the other logic forms described. It uses m.o.s.f.e.ts in place of transistors and resistors, resulting in a very small chip size and making it possible to have more functions per integrated circuit. This means that the main area of application of m.o.s. is where there is a requirement for a large number of logic functions on a single chip.

The basic gate circuits are shown in Figs. 7 and 8. In the circuit shown in Fig. 7, the output goes positive only if both inputs are taken negative, i.e. Tr_1 and Tr_2 are turned on.

In the circuit shown in Fig. 8, the output







Fig. 5. t.t.l. gate with Tr_4 switched on. The V_{be} and V_{ce} under saturated conditions are assumed to be 0.6 V and 0.1 V respectively. From the voltages shown in the diagram it is easy to see that the base voltage necessary to turn Tr_3 on is 1.3 V thus ensuring that it is turned off. However if D_1 were omitted, there is every likelihood of Tr_3 being turned on.



Fig. 6. Basic e.c.l. gate.



Fig. 7. m.o.s. NAND gate.



ig. 8. m.o.s. NOR gate.

oes negative only if both inputs are taken ositive. i.e. Tr_1 and Tr_2 are turned off. **n** both circuits the load "resistor" is Tr_3 thich is designed to give a suitable resist--nce value by controlling its g_m . The lajority of m.o.s. logic circuits use only -channel devices but some circuits using - and n-channel devices are now becoming vailable. The supply voltages for these rcuits is of the order of 20 V and the logic wing is of similar proportions. The beeds of operation are relatively slow, -ropagation delays being in the region of 00 to 500 ns.

lossary of terms and abbreviations

he list which follows explains the initials sed to describe most of the logic families.

- .c.s.l. -compatible current sinking logic 11. -counting logic
- -complementary transistor logic t.l.
- .t.l. -diode-transistor logic
- c.c.s.l. -emitter-coupled current steered logic
- c.1. -emitter-coupled logic
- e.c.l. or
- ².c.l. -emitter-emitter-coupled logic
- -high noise immunity logic .n.i.l.
- .l.t.t.l. --high-level transistor-transistor
- logic
- -high threshold logic .t.l.
- -metal-oxide-silicon (logic) 1.0.S. c.t.l.
- -resistor-capacitor-transistor logic -resistor-transistor logic
- t.1. t.l. or
- .1. -transistor-transistor logic

Some terms used in connection with itegrated logic circuits are as follows:

an-in This is the maximum number of iput signals which may be fed into a gate. the gate has four input points the "fan-in" said to be four. Note: In circuits where all ie available inputs are not used, it is good actice to connect the unused input points) that they cannot affect the operation of the te. As a general rule, with r.t.l., unused put points should be connected to the egative rail.

-an-out This is the maximum number of ads (gate inputs) which the output of a rcuit is capable of driving. It is normally loted as a simple number. (The effective umber of loads represented by inputs of tegrated circuits are usually quoted as well they are not necessarily unity). Fan-out

is also a fair indication of power output capability as it is usually possible, from a knowledge of what one load is equivalent to, to calculate the available output power.

Noise margin This can be defined as the interfering input voltage necessary to cause the gate to change state or start to change state when the input is in either of its normal operating conditions. There are two switching thresholds. (See Fig. 9). One is associated with the turn on and the other with the turn off. The noise margin, quite clearly, is the difference between the specified "o" and "I" states and the respective switching thresholds. The noise margins are not necessarily the same and are usually quoted accordingly. In some e.c.l. gates there is also a further noise margin in that, if the "I" voltage is increased too far, part of the circuit will saturate and a phase reversal will occur as shown dotted.

Power dissipation This is usually expressed as an average value with the gates operating with a 50% duty cycle. However, some manufactures quote dissipation figures for both the on and off conditions. With saturating logic it is quite usual for the power consumption to increase with frequency.

Propagation delay This is a measure of the speed of operation of a gate. Fig. 10 shows that there are propagation delays associated both with turn-on and turn-off times. The turn-on delay is usually shorter than turn-off delay but this is by no means a fixed rule. However, because the delays are frequently different, manufacturers so usually quote average delay. This is one of the reasons for the very wide propagation delay tolerance in many data sheets.



Fig. 9. Diagrammatic representation of the transfer characteristic of a gate.



Fig. 10. Waveform showing propagation delay using a non-inverting gate. DI is the "turn on" delay and D2 the "turn-off" time. For convenience it is assumed, in this illustration, that the switching point is at the centre of the transition from 0 to 1. In practice this is very seldom so.

May Meetings

Tickets are required for some meetings: readers are advised, therefore to communicate with the society concerned

LONDON

1st. R.T.S .- Symposium on "Control room design and layout" at 17.00 at the I.T.A., 70 Brompton Rd., S.W.3.

2nd. I.P.P.S .- "Acousto-electric effects in semiconductors" at 10.30 at Imperial College, S.W.7.

5th. I.P.P.S.-"Organization and management of research and development" at 10.15 at I.E.E., Savoy Pl., W.C.2.

- 5th. I.E.E.—"V.L.F. navigation" by S.S.D. Jones at 17.30 at Savoy Pl., W.C.2.
- 6th. I.E.E. & I.Mech.E .- "Electronic turbine governing" by P. A. L. Ham and A. A. L. Bental at 17.30 at

Savoy Pl., W.C.2. 7th. I.E.R.E.—"The Rapier ground-to-air missile sys-tem" by S.C. Dunn at 18.00 at 9 Bedford Sq., W.C.1. 12th I.E.E.—"Filters with periodically time-varying parameters" by Dr. W. Saraga at 17.30 at Savoy Pl., W.C.2.

14th. I.E.E.-"Large scale integration-why, where and when?" by D. D. Jones at 17.30 at Savoy Pl., W.C.2.

14th. S.E.R.T .- "Applications of the unijunction transistor" by G. C. Rayworth at 19.00 at the London

School of Hygiene, Keppel St., W.C.1. 15th. I.P.P.S.—Symposium on "Optical techniques in acoustics" at 14.30 at the Physics Dept., Imperial College, S.W.7.

F. J. Wilkins and M. J. Swann at 17.30 at Savoy Pl., W.C.2.

15th. I.E.R.E .- "Absolute digital displacement transducers" by A. L. Whitwell at 18.00 at 9 Bedford Sq., W.C.1.

19th. I.E.E .- Colloquium on "Cathode-ray tube display" at 10.00 at Savoy Pl., W.C.2.

20th. I.E.E. & I.E.R.E .- "Peripheral auditory mechanisms" by Dr. H. A. Beagley at 17.30 at St. Bartho-

lomew's Hospital, E.C.1. 21st. S.E.R.T.—"Practical work in technician courses" by A. J. Hymans at 19.00 at the Royal Society of Arts, John Adam St., W.C.2.

CAMBORNE

15th. I.E.E.-"The operational aspect of Eurovision" by A. R. Elliott at 14.30 at Cornwall Technical College.

CARDIFF

16th. S.E.R.T .- "Tape recorder servicing" by H. W. Hellyer at 19.30 at the Llandaff Technical College.

COLCHESTER

15TH. I.E.E.-"Tomorrow's world in tele-communications" by W. J. Bray at 18.30 at Essex University.

HORNCHURCH

8th. I.E.R.E.-"Automatic production testing of electronic equipment" by R. Kitchen at 18.30 at the Technical College, 42 Ardleigh Green Road.

MANCHESTER

5th. I.E.E.—"Invention as part of education" by Prof. M. W. Thring at 18.15 at U.M.I.S.T.

NEWCASTLE-UPON-TYNE

7th. S.E.R.T.—"Colour receiver decoder and c.d.a. circuits" by L. H. Briggs at 19.00 at the Charles Trevelyan Technical College, Maple Terrace.

PLYMOUTH

7th. R.T.S.-"'Translators and repeater stations" by D. L. Smari at 19.30 at Westward TV Studios.

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Personalities

Sir Ian Orr-Ewing, Bart, O.B.E., M.A., F.I.E.E., M.P., is the new president of the Electronic Engineering Association. Sir Ian, who is 57, is chairman of Ultra Electric (Holdings) and of Ultra Electronics which he joined in 1966. After graduating from Trinity College, Oxford, with an honours degree in physics he served a three-year apprenticeship with E.M.I. and then joined the B.B.C. Television Service in 1937 where he stayed (except for war service in the R.A.F.) until 1949. He then served as a director of the Cossor Group of Companies until 1957 when he resigned on being appointed a minister in the Government. Sir Ian served as Un-



Sir Ian Orr-Ewing

der Secretary of State for Air (1957-59) and Civil Lord of the Admiralty (1959-63). At on time during the war he was chief radar officer on General Eisenhower's staff.

Among those recently elected Fellows of the Royal Society were:— **Professor R. L. F. Boyd, Ph.D.,** B.Sc., F.I.E.E., professor of physics and director of the Mullard Space Science Research Laboratory, at University College, in the University of London, "distinguished for his contributions to ionospheric physics and X-ray and ultraviolet astronomy and to the exploitation of space science techniques in these fields"; **Professor C. W. Oatley**, O.B.E., M.A., M.Sc., F.I.E.E., professor of electrical engineering at the Department of Engineering in the University of Cambridge, "distinguished for his work on measurements at microwave frequencies, in electron-optics, and in particular for the design and development of a successful scanning electron microscope"; and Dr. J. H. Wilkinson, deputy chief scientific officer, Mathematics Division, National Physical Laboratory, Teddington, "distinguished for his contributions to numerical analysis and the development of digital computers, and especially for his work on backward error analysis".

Elizabeth Laverick, Ph.D., B.Sc., F.I.E.E., A.Inst.P., head of research and advanced projects at Elliott-Automation Radar Systems Ltd, has been awarded an honorary fellowship of the University of Manchester Institute of Science & Technology in recognition of her achievements in technology and technological education. Dr. Laverick, who has made major contributions in the field of radar and microwave engineering, gained her B.Sc. at Durham University where she later became the Physics Department's first lady Ph.D. She then joined the microwave aerials department of the GEC Applied Electronics Laboratory. She joined Elliotts in 1954 as a microwave engineer and, in 1959, became head of the Radar Research Laboratory. Since 1967 Dr Laverick has been a member of the Electronics Divisional Board of the Institution of Electrical Engineers, the first lady member, and is president of the Women's Engineering Society.

E. Thompson, B.Eng., M.I.E.E., and V. J. Cox, M.B.E., have been appointed directors of Ekco Electronics Ltd. Joining Ekco in 1949, Mr. Thompson was appointed head of nucleonic development four years later. In 1963 he became technical sales manager-instrumentation, and in 1967 took charge of the Instrumentation Division and will continue to do so: Mr. Cox joined Ekco in 1941 and for many years he has been wholly responsible for avionic design matters. In 1959 he was appointed chief engineer-aviation, and in September last year he became manager of the Aviation Division, a position which he will retain.

John H. Buying, aged 46, who originally joined Marconi Instruments as an X-ray development engineer in 1948, following service with Philips Electrical Ltd, has been appointed sales manager of the Sanders Division of the company. In 1950 he became a service engineer with General Radiological Ltd. In 1954 he rejoined M.I. as a sales engineer in the Export Department, and he was later appointed a distributor manager. The company also announces the appointment of Renie G. Weston as sales engineering supervisor. Following service in ground radar in the R.A.F. during the last war, Mr. Weston, who is 46, became a sales engineer with Pye Telecommunications Ltd in 1946. He subsequently became European sales manager. From 1960 to 1964 he was sales director of Storno Southern Ltd, and in 1965 he joined the Specialised Components Division of the Marconi Company, transferring to M.I's Sanders Division last October.

John C. W. McCarthy, B.E.M., has joined Racal Communications Ltd as systems consultant. He has spent over 40 years in the Civil Service for 33 of which he was concerned with electronics in the Royal Naval Scientific Service, latterly with the Ministry of Defence (Navy).

R. M. Carroll is now managing director of Eddystone Radio, a member of GEC-Marconi Electronics, on the retirement of H. Cox who had been with the company since its formation in 1927. Richard Carroll, who is 45, joined the Marconi Company's test department at Chelmsford in 1947. He was in the supplies department from 1958 until 1964 when he was appointed marketing manager of the newly formed Microelectronics Division of the company. He was transferred to Eddystone as works manager about a year ago. Kenneth R. Williams, who joined Eddystone in 1938, is appointed sales manager. After being in charge of the service department for seven years he took over the professional equipment sales department in 1945.



R. M. Carroll

Wireless World, May 196

P. E. Leventhall, B.Sc. (Hons F.LE.R.E., has been appointe division manager of S.T.C's Marin Division at Croydon (Internationa Marine Radio Company Ltd). H had previously been technical mar ager for I.M.R.C. and Hudso Electronics. Mr. Leventhall, who 41, joined I.M.R.C. in 1966 fron Cossor Communications Ltd, when he had been chief engineer.

Dennis C. Flack, Ph.D., B.Sc F.I.E.E., has been appointed chinengineer of Sifam Electrical Instrment Co. Ltd, of Torquay. D Flack, who is 44, was with Sangar Controls, a division of Sangar Weston Ltd, from 1960 to 196. Prior to this he was for twenty yea in the aircraft industry at Briste latterly as chief electrical design with Bristol Aircraft Company.

GEC-Marconi Electronics Ltd a nounces the formation of a ne Mobile Communications Divisit based at Spon Street Works, Cove try, the manager of which is I. Alexander, B.Sc., F.I.E.E., forme ly the technical director of the Comunications Division of GEC-A (Electronics) Ltd. J. E. Hills is apointed sales and marketing ma ager, and D. A. S. Dryboroug B.Sc., M.I.E.E., chief engineer.

Robin Stephens, who joined Wi Electronics Ltd, of Bognor Reg twelve months ago as marketi manager has been appointed to t board as marketing director. graduated in electrical engineeri from Bristol University and af serving as sales manager in t Computer Division of Solartr Electronic Group, was later mark ing manager of Redifon Astroda Ltd.

K. G. Thorne, F.I.E.E., F.I.E.R. has joined Epsylon Industries L of Feltham, Middx, as chairman a managing director in succession I. D. Cuffe who has taken up corporate appointment with the Hsylon parent organization, Leigh struments, of Ontario, Canada. M Thorne was formerly managi director of Computing Devices (Ltd, London.

OBITUARY

John Clarricoats, O.B.E., who has c tributed our "World of Amateur Radsection since its introduction in 19 died on March 7th aged 71. "Cları as he was affectionately known among amateur transmitting fraternity, star his radio career with Standard T phones & Cables Ltd. and bec: full-time secretary of the Radio Soc of Great Britain in 1932. He reti from this post in 1963 but contin as honorary secretary of the Europ Regional Division of the Internatic Amateur Radio Union. John Cla coats also played a major role in le government (particularly in educatic matters) and was Major of the Lom-Borough of Enfield. I recolores intered testing either involutike you can take the components straight from the pcstmar to the assembly line! Electrosil quality is the same however small the resistor.

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G7

Wireless World Colour Television Receiver

12. Chrominance circuits

Last month details were given of one of the two main boards of the decoder, the board which carries the input chrominance amplifier, the burst amplifier and burst gate, the phase discriminator, the crystal oscillator, the colour killer and the identity circuits. This month the other main board is being dealt with and Fig.1 shows its circuit diagram. As before, the numbers in brackets refer to the sections of the block diagram of Part 10.

The input is applied at P_{12} , which is joined to P_2 in the other board. This signal is the complete chrominance signal plus colour burst. It is applied through C_{35} to a combined attenuator and burst gate (3), the purpose of the gate being this time to remove the burst from the signal.

Saturation control

The attenuator comprises two diodes D_9 and D_{11} connected back to back. Their cathodes are joined to the chassis through the resistors R_{50} and R_{52} , while a variable bias is applied to their anodes. The resistance of the diodes varies with this bias and, therefore, the amount of signal transmitted through them. One resistor, R_{51} , is returned to -20V. Another, R_{57} , is taken through R_{58} to the slider of R_{60} to which up to 15V can be applied. This control R_{60} is not mounted on the board, but on the front panel, and it forms the saturation control. It is a purely d.c. control and so can safely be connected by leads of any length.

The junction of R_{57} and R_{58} is taken via P_{14} to the luminance amplifier where it is linked with the contrast control. It is connected to the point marked "To saturation control in chrominance units" in Fig.1, Part 6. Adjustment of contrast varies slightly the bias applied to the attenuator diodes and tends to make the saturation independent of brightness.

Another diode D_{10} is connected to the anodes of D_9 and D_{11} . A negative-going line flyback pulse of some 80V amplitude from the line timebase is fed to P_{18} by a screened cable. This is reduced to about 3.5V by the potential divider R_{63} , R_{62} and applied through R_{61} to the cathode of D_{10} . It renders D_{10} conductive and makes the anodes of D_9 and D_{11} negative, cutting them off. The colour burst, which occurs during the line flyback, is thus prevented from passing through the diodes and the attenuator output at P_{13} is the chrominance signal only at an amplitude which depends on the setting of R_{60} .

This signal is applied to the chrominance amplifier (4) Tr_9 which has a tuned base circuit comprising L_6 and C_{39} damped by R_{53} and tuned to 4.43 MHz. It has the usual emitter bias network, but the base is taken via P_{20} and P_8 in the other board to the anode of D_8 . When there is no colour signal no bias is applied and Tr_9 is cut-off, thus preventing noise or interference from passing any further through the chrominance circuits to appear on a monochrome picture. With a colour signal a positive bias is developed by D_8 in the way explained last month and it appears at P_{20} and renders Tr_9 operative.

The amplified chrominance signal at the collector is passed to the next stage Tr_{10} through a trap L_7 , C_{45} tuned to 6MHz. Tr_{10} is connected as a phase-splitter (5) with both collector and emitter loads. The collector load is the 150- Ω resistor R_{72} which is connected to the input of the PAL delay line (6) via P_{21} and P_{22} .

At signal frequency the emitter load comprises a fixed component of 86Ω provided by R_{76} and R_{77} in series, shunted by a variable component of 39Ω to 139Ω provided by R_{74} and R_{75} in series. The variable control R_{75} enables the relative amplitudes of the emitter and collector outputs to be adjusted. At d.c. the emitter resistance varies from 159Ω to 259Ω as R_{75} is adjusted, but this does not vary the operating point of the transistor seriously.

The emitter output is taken from the junction of R_{76} and R_{77} to the centre point of the 1:1 auto-transformer T_7 (7). The two ends of this winding are taken, one each, to the two synchronous demodulators. The centre-point and the R-Y end are taken to P_{27} and P_{28} and thence to the output of the PAL delay line. The 150- Ω resistor R_{95} produces the proper termination for the line.

Because of the auto-transformer action the signal delayed by the PAL line appears at equal amplitudes but opposite phase at the two ends of the auto-transformer. At the one end it is added to, and at the other end it is subtracted from, the undelayed signal applied at the centre tap. Two things happen as a result. One is that the chrominance signal is separated into its R-Yand B-Y components; the demodulators alone are capable of doing this, and in simple PAL they do, but this pre-separation renders them less critical. The other and main thing is that by combining the signals of one line with those of a previous line, delayed in time by one line, phase errors are largely cancelled.

Reference oscillator inputs

At this point it is necessary to stop following the path of the chrominance signal for a while, because the demodulators require inputs derived from the reference oscillator. For a proper understanding of the demodulators it is necessary to see how these are derived. The output of the reference oscillator at sub-carrier frequency is taken from P_6 in the other board through a $0.002-\mu$ F capacitor to P_{26} in this board. Notice particularly that this capacitor does not appear in the diagrams because it is connected directly between P_6 and P_{26} and so is not mounted on either board.

The transformer T_4 has four identical windings, and the reference oscillator output is fed to one of them. This winding and one other form a 1:1 ratio auto-transformer and across the whole the voltage is double the oscillator input and is balanced to chassis. This is applied to the phase-shifting (36) circuit C_{57} ,

 R_{90} . When the resistance equals the reactance of the capacitance the voltage appearing between their junction and chassis is equal in amplitude to the voltage fed in from the oscillator, but is shifted 90° in phase.

The primary of another transformer T_6 is connected between these points and its secondary feeds a voltage at the frequency of the reference oscillator, but at 90° in phase to it, through R_{93} and R_{94} to the B-Y demodulator.

Returning to T_4 , two secondaries are connected as shown to D_{15} and D_{16} which are arranged to be conductive alternately. When D_{15} is conductive its secondary feeds the oscillator frequency through C_{54} , the primary of T_5 and D_{15} . When D_{16} is conductive the secondary connected to it feeds the primary of T_5 through C_{55} and D_{16} . The outputs of the two secondaries are in opposite phase and so the phase of the voltage fed to T_5 reverses whenever the diodes are switched (35).

PAL switch

This switching is effected by the bistable (34) Tr_{11} and Tr_{12} . This is conventional and needs little or no explanation. The transistors are cut-off and conductive alternately. When one is conductive its collector is nearly at chassis potential and the collector of the other, which is then cut-off, is at about 15V. The collectors are joined through R_{86} and R_{87} to the two diodes D_{15} and D_{16} .

If Tr_{12} is cut-off, for example, the cathode of D_{15} is connected through R_{86} and R_{83} to 15V. The anodes of both diodes are taken to the potential divider R_{88} , R_{89} at about 10V. There is thus about 5V reverse bias applied to D_{15} through some $2k \alpha$. The other diode D_{16} has its anode at about 10V but its cathode is nearly at chassis potential. It thus has nearly 10V applied in the conductive direction. When the bistable changes state the conditions are reversed and D_{15} conducts while D_{16} cuts-off.

The change of state of the bistable is carried out at line frequency by a pulse derived from the line timebase (33). The -80-V pulse occurring on the line flyback at P_{18} is applied through C_{42} shunted by R_{64} to D_{12} and renders it conductive. This diode is normally cut-off by 15V reverse bias applied through R_{66} . The pulse current flows through D_{12} mainly into C_{43} charging it negatively to chassis. When the diode becomes non-conductive again, which it does very shortly after the peak of the flyback pulse, and so about half way through the flyback period, the tuned circuit L_5 , C_{43} , R_{65} is left isolated with C_{43} charged negatively.

It must be pointed out that because of the heavy damping provided by R_{65} , C_{43} does not charge very quickly as one might expect. The voltage waveform during the charging period is nearly a quarter cycle of sinewave.

On the discharge the waveform is a damped sinusoid, which starts at the negative peak and dies away in some two or three cycles. Because of the near sinusoidal voltage during the charging time, however, the charging and discharging parts of the waveform are not really separable when the wave is viewed on an oscilloscope. The waveform looks like a damped sinewave which starts with a negative half cycle of about 5V amplitude and is followed by a positive half cycle of some 3.5V amplitude. Successive half cycles die away rapidly.

It is the first positive half cycle which is used and the circuit has its frequency adjusted by L_5 so that its peak coincides with the centre of the colour burst. The wave is applied through P_{25} to P_3 in the other board where it renders Tr_3 conductive during the burst. It is also applied to the junction of C_{50} and C_{51} and here it is the first negative half cycle which is important.

Fig. 1. Circuit diagram of the second main board of the decoder. It includes the chrominance amplifier, the PAL switch, the synchronous detectors and the first-stage video amplifier and matrix circuit



It renders both D_{13} and D_{14} conductive and so carries the basis of both transistors negatively. This has no effect on the transistor which is not conducting, but it reduces the current in the conducting transistor and because of the cross-coupling between the two this starts to turn on the other. The normal regenerative action then causes the bistable to change state. Thus the bistable changes state once every line during the flyback period and this reverses the phase of the oscillator signal applied to T_5 once every line in step with the R-Y reversals of phase in the signal

While the bistable will change state every line and so reverse he phase of the oscillator signal on T_5 once every line, there is no guarantee that the phase will be in its proper relation with he signal; it may always be 180° out of its proper phase. To prevent this an identity circuit is provided. This was explained ast month, when it was shown how a roughly sinusoidal voltage of half line frequency (7.8kHz) is developed and taken through D_6 to P_7 . From there it is brought in to either P_{23} or P_{24} .

Suppose that it is taken to P_{24} . If it happens that Tr_{12} is conductive, the collector potential is very low and so is the potential of P_{24} . If Tr_{12} is non-conductive, however, the collector potential is nearly 15V, D_{14} is cut-off and so is D_6 and the positive half cycle from Tr_8 is not passed by D_6 and the positive pulse from L_5 acts normally on Tr_{12} to initiate a change of state. During the next line Tr_8 produces a negative half cycle which is not passed by D_6 . Thus the identity circuit does nothing.

However, if Tr_{12} is conductive when the positive half cycle of identity signal occurs matters are different. The potential of P_{24} is then near earth, D_6 conducts and allows the positive half cycle to reach P_{24} . It is now Tr_{11} which is conductive and he positive pulse from L_5 makes it draw current as usual. The positive half cycle at P_{24} now holds the base of Tr_{12} positive and prevents it from moving negatively in response to the change in Tr_{11} . It thus prevents the usual change of state from occurring. It must be noticed that D_{14} is conductive because its anode is held positive by its connection to R_{81} and R_{85} .

The result is that if it happens that the bistable is being triggered in the wrong phase, the identity signal prevents it from triggering, and it remains in the same state for two consecutive lines, and this brings it into the right phase.

Two points of connection, P_{23} and P_{24} , are provided for the identity signal but, of course, only one is used. The correct phase of the bistable depends on the phasing of some of the transformers. If a mistake is made in one of these, the identity circuit will keep the bistable consistently in the wrong phase. It is usually easier to correct such an error by transferring the identity lead from one pin to another than to find the transformer with the wrong connections and reverse its leads.

This must not be taken to mean that care over the transformer connections is unnecessary. Not all errors of connection can be remedied by changing over the identity lead.

It will now be clear how the reference oscillator signal, reversing in phase every line, is obtained across T_5 . The secondary output is applied to the R-Y demodulator through the resistors R_{91} and R_{92} . A tuned circuit is connected across the secondary, however, in order to suppress harmonics. This comprises L_8 tuned by the combination of C_{68} to C_{70} .

Synchronous detectors

All four diodes of a demodulator conduct during the positive half cycles of the reference signal and so effectively join the input and output terminals and allow the chrominance signals to pass. During the negative half cycles all four diodes are nonconductive and the input and output terminals are isolated.



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The outputs are in the form of half cycles of the 4.43-MHz signal waveform and these are smoothed by C_{59} and C_{61} to the R-Y and B-Y video signals. Trap circuits L_9 , C_{58} and L_{10} , C_{60} tuned to 4.43 MHz, remove this component from the signals. The output loads, which are also the base resistors of the following amplifiers, are R_{96} and R_{97} .

Colour difference amplifiers

The chrominance signals now exist in video form and it is necessary to amplify them and also to produce from the R-Yand B-Y signals the G-Y signal. The red (12) and blue (13) channel amplifiers are substantially the same, and are Tr_{13} and Tr_{15} respectively. In each case there is an emitter load of 22k ρ returned to -20V. The collector loads are taken to 20V, so the transistors operate with a total supply of 40V. The load of Tr_{13} is a fixed resistor R_{104} of $4.7k \rho$, but that of Tr_{15} is variable from $3.3k\rho$ to $8.3k\rho$ by R_{108} . The blue channel requires higher gain than the red and it must be adjusted to be the right proportion to it.

The emitters have a.c. loads which are virtually provided by R_{105} of 1.2k and R_{110} of 680. This alone makes the red channel have nearly twice the feedback of the blue channel and, hence, makes the gain of the latter nearly double. The emitters of these two transistors are joined through R_{99} and R_{100} shunted by R_{98} , which permits the precise ratio of these resistances to be adjusted. The combination of the R-Yand B-Y signals so obtained is applied through C_{62} to the emitter of Tr_{14} and is the G-Y signal. This transistor operates under the same conditions as the other two, and has an adjustable collector load R_{107} , R_{106} to enable the G-Y signal to be set at the proper level.

From the collectors of these three transistors onwards the three channels are identical, so it suffices to describe one of them. The coupling capacitors and grid leaks of the following valves (16)-(21) are mounted on the board and are terminated at P_{35} , P_{36} and P_{37} . The grid stoppers R_{111} , etc., are connected directly between these pins and the grid terminals of the valveholders.

All the parts shown in Fig.1 are mounted on a strip of Veroboard of the same dimensions as that used for the reference oscillator, etc., described last month. Fig.2 shows photographs of both sides of this board.

Details of all the transformers for both boards were given last month. Details of the coils used in both will be given next month. With the exception of L_4 , all the coils are wound on the same type of former and are of the kind used in the i.f. unit. The bases have six pins, of which only two are used for connections. The pin spacing does not fit that of the holes in the

Fig. 2. Photographs of the two sides of the board, showing the component layout. Note that C_{56} does not appear on the board because there was just enough capacitance in C_{57} without it. In some cases, it may be required, but will rarely need to be more than 10pF

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Earth tag

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Wireless World, May 1969

Veroboard and the board cannot be re-drilled for them because of the existing holes. The best thing to do is to cut off the middle pair of pins from the coil bases, leaving only the outer four in the form of a square. Four existing Veroboard holes can then be enlarged with a drill so that the coil base is a tight fit into the four holes, although the individual holes are really very large for the pins. This can just be done without breaking the copper strips, and all four pins are then soldered to the strips, which are broken in the appropriate places. With so many components as in these boards, it is virtually impossible to bring the coil pins just where one would like, and wire connections for the coils are often needed.

The coil L_4 is on a larger former with a larger can. The can is mounted on its side by a pair of self-tapping screws into the side of the can. A small piece of Veroboard has the coil mounted on it and the coil connections made to it, and is held to the fixing lugs of the can by two 6 B.A. screws and nuts. The two tuning capacitors for this coil are mounted on it in the can.

The corner hole of each board is enlarged to 6 B.A. clearance for mounting. Each adjacent hole in the board has the copper removed so that the earthed mounting screws and nuts do not introduce any short-circuits.

The whole of the decoder circuits are mounted in a frame built of brass angle. The horizontal flanges on the board side at the input end each have a $\frac{1}{4}$ -inch hole, through the two of which is passed a length of $\frac{1}{4}$ -inch rod which is fixed vertically to the base. The bottom flange rests on the base, and the rod is sufficient to hold the assembly securely and yet enables it to be swung outwards to the rear for access while in operation.

The boards are fixed to the frame by long 6 B.A. screws and nuts. The PAL delay line is attached to a separate board which is screwed to the back of the frame. This is a piece of plain Veroboard with a 0.1-inch matrix of holes; this is necessary to fit the tags and mounting pins of the delay line but, of course, plain board can be used if it is drilled appropriately.

The connections for the delay line are soldered to P_{21} , P_{22} , P_{27} and P_{28} and taken through enlarged adjacent holes to pins on the delay-line side of that board. On the other side the delay line pins are themselves joined to the board pins, and a chassis earth lead is taken to the frame of the line.

In this way the whole delay line with its board is readily disconnected and removed if access is required to the back of either of the main boards. Access at the back is fairly good except at the top of the upper board and the bottom of the lower, where the boards are screened by the flanges of the framework.

The equipment can be operated in the simple PAL mode merely by disconnecting the delay line completely; that is, by making no connection to P_{21} , P_{22} , P_{27} and P_{28} . A few other components are then unnecessary, but their presence does no harm. There are certain advantages in doing this during the initial alignment. This together with details of the output stages will be given in Part 13.





+20V

Physics Exhibition

Selected items seen at Alexandra Palace, London, March 10-13

Ultrasonic holography

An experimental equipment for making holograms at ultrasonic frequencies was demonstrated on the E.M.I. stand. The demonstration was carried out underwater at 5MHz, although frequencies between 500kHz and 10MHz would have been suitable under these conditions.

The object to be viewed was suspended in a tank of water in the path of a beam from an ultrasonic transducer. Reflected energy from the object was combined with the output of a second transducer which was used to supply the reference signal. The interference pattern so formed was then mechanically scanned, a receiving transducer being used to sense the interference pattern.

The output of the receiving transducer, after suitable amplification, intensity modulated an oscilloscope which was made to scan in sympathy with the receiving transducer. The information on the oscilloscope tube face was then photographically recorded. Instead of being presented on an oscilloscope the output of the receiving transducer could have been used to intensity modulate a light source.

The hologram is viewed in the normal way, the film being illuminated with the output of a laser. On the stand the viewing was done using a closed-circuit television system in order to remove the hazards of viewing laser light with the naked eye.

The resolution obtained using this technique depends upon accurate synchronization, the wavelength used and the size of the scanned area. Unwanted reflections from the side of the tank can be virtually eliminated using suitable gating; however, there is still some break-through of the reference signal. This can be



Ultrasonic holography demonstration apparatus showing, from left to right inside the tank: scanning transducer; transmitting transducer; target; reference transducer. (E.M.I.)

defocused to some extent by critical positioning of the hologram when viewing.

E.M.I. suggest that a possible use for the technique would be to view objects under difficult conditions—in fog, in muddy water or under skin tissue, for instance.

Voice-operated typewriter

Designed as an aid for the disabled, a voice-operated typewriter developed by Standard Telecommunication Laboratories uses a simple electronic recognizer which responds to messages spoken in a modied version of the Morse code. The sounds 'di' and 'dah' (representing the dot and dash of the code) are used, and the machine recognizes these not on a spectral or phonemic basis but purely by their different durations. The spoken message is picked up by a microphone, and the output of this is band-pass filtered and fed to a threshold detector which detects the presence and defines the duration of peaks of voicing in the speech waveform corresponding to the 'di' and the 'dah'. There follow two further duration threshold devices (monostable circuits), one to differentiate between short and long sounds ('di' and 'dah') and the other to discriminate between silences within a code group (letter) and silences separating code groups. The 'di' and



Voice-operated typewriter, showing electronic recognizer on the left. (Standard Telecommunication Laboratories)

'dah' information is then translated into a binary code ('di'=0, 'dah' = 1) and progressively fills a four-stage shift register (four bistable circuits) as the sounds are uttered. For example, for 'n' (dah-di), the register is altered from its basic state of 0000 to 0100. In addition, an associated two-stage counter (two bistable circuits) counts in binary notation the number of sounds uttered —for 'n' the count is 10. The letter 'n' is thus uniquely identified by the six-digit code 010010.

When the duration threshold device recognizes the end of the uttered group (a long silence) the contents of the shift register and counter are read into a decoder, the output of which actuates the corresponding typewriter key. It is claimed that a person can operate a typewriter accurately by this method at up to 20 words a minute, after a few hours' training. The output of the recognizer can, of course, be used for other functions such as 'dialling' telephone numbers or switching domestic equipment on and off.

Acoustic parametric receiving array

A highly-directional acoustic receiving array, covering a wide bandwidth of response, has been developed at Birmingham University. The system uses the non-linear interaction of a signal wave with a powerful local 'pump' wave.

The 'pump' transducer produces a high-frequency (about 6MHz) acoustic wave, which is well collimated within the Fresnel diffraction region, and a transducer placed along the axis of the first acts as a receiving probe. First-order interaction
between an incoming acoustic plane wave and the locally generated 'pump' column causes sum and difference frequency components to be generated in the water. The magnitude of the pressures at the receiving probe at these interaction frequencies can be shown to depend on the angle between the incoming wave and the pump wave. If the signal frequency is much less than the pump frequency, the device behaves very much like an ordinary end-fire array of the same length operating at the signal frequency.

In the interaction process, energy is transferred from the pump wave into the new frequencies, producing an up-converter type of parametric amplification. This means that a low-level acoustic signal is at least maintained at the same intensity on arrival at the receiving probe (at the new frequencies).

In this way it is possible to make a wide-band receiving device with acceptable directivity even at very low frequencies using two small transducers only.

Electromechanical resonator for i.cs

A high-Q resonator that could possibly be used for filters, tuned circuits or oscillators in hybrid integrated circuits is based on the mass and compliance of a small mechanical element. The unusual feature is the use of electrostatic transducers for driving and pick-up. Shown by Standard Telecommunication Laboratories, the device consists basically of a metal beam mounted over conducting plates which form the input and output electrostatic transducers. The variable electrostatic force caused by the incoming signal acts on the metal beam and excites it into flexural vibration corresponding to its natural frequency. This frequency depends on the beam material and on the physical dimensions. The vibration causes small variations in the capacitance of the output transducer which are detected with a high input impedance amplifier.' The device is completely passive and is reversible; by variation of the coupling it can be made symmetrical or unsymmetrical. Frequency range is from a few hundred hertz to a few hundred kilohertz, but the most suitable range is 1-20kHz. At low frequencies, below 1kHz, the effect of unwanted external vibrations can be reduced by using a balanced resonant element. S.T.L. say there is considerable choice in the shape and material of the resonant element and in the arrangement of the conducting plates, so a wide range of application should be possible. The vibrating beam can be mass loaded and tuned by removing material from the loaded end. Fine tuning can also be carried out by variation of bias voltage.

Microwave integrated circuits

Applications of microwave i.cs (microstrip) incorporating p-i-n diodes, shown by AEI Semiconductors, included a working circuit of a sideband generator in which two p-i-n phase shifters were included in a ring hybrid. The diodes were modulated in push-pull at a variable frequency up to a few MHz and each changed the phase of the signal by 180°. The fundamental cancelled out and sidebands corresponding to odd harmonics of the modulation frequency were left. Other microstrip circuits shown were a phase-shifting shunt mounted diode, a low capacitance series-mounted diode, a broadband changeover switch and broadband a.m. modulator. Our photograph illustrates a 180° phase-changer using two p-i-n diodes and two chip capacitors for d.c. blocking. The use of integration techniques makes for a tremendous reduction in size and cost of microwave equipment, the switching facility being particularly useful because of the practical difficulties of mechanical switching at microwavelengths.

A practical application of microwave i.cs could be seen in a display by the Admiralty Surface Weapons Establishment, where a microstrip pulsed Gunn oscillator was incorporated in



Microstrip 180° phase-changer incorporating two p-i-n diodes. (A.E.I. Semiconductors Ltd.)

an X-band marine radar beacon. The whole system, aerial, transmitter and receiver, occupied a p.c. board measuring about 152×100 mm. The transmitter has a peak output power of 100mW and when the radar beacon, or "Racon" as it is called, is fitted to a buoy, it transmits a long identifying pulse on receipt of an interrogating pulse from a ship's radar. The pulse then marks the position of the buoy on the ship's radar. Circuits based on the microstrip transmission line are still being developed, using both thin and thick film techniques.

Motor using piezo-electric effect

A simple reversible linear motor has been designed at the Royal Radar Establishment. This motor moves in steps along a precision track, and the size of step can be varied within the range $0.1-4.0\mu$ m, with final adjustments to 0.01μ m. The basic structure is very simple. Two electro-magnets, with pole pieces in contact with the horizontal track, are separated by a length of ceramic tube. A voltage applied between the inner and outer surfaces of the tube results, by piezo-electric effect, in a shortening of the tube by an amount proportional to the applied voltage. A single step is taken by de-energizing one magnet to free it from the track, shortening the ceramic tube as described, de-energizing the other magnet, clamping the first magnet, removing the contracting voltage from the tube (causing the free pole pieces to slide along the track), and finally reclamping the remaining magnet.

Integrated circuits are used in a simple logic unit which controls the above cycle and 5V signals are used to start and reverse the motor by remote control. A variable voltage in the range 0-9V determines the step size, by controlling the h.t. generator, and the stepping rate may be controlled by an external oscillator if required.

The distance over which the motor will travel is limited only by the length of precision 'V'-block used as track, and the length of flying leads provided. The ceramic is available in a variety of sizes, a factor enabling motor size and performance to be varied.

Variations in local force will change the size of step obtained for a given voltage, so that for some applications it would be necessary to use an optical or other highly accurate measuring technique to take full advantage of the motor's capabilities. Possible uses for the motor include: (1) general micromanipulation, (2) moving specimens in microscope work, (3) preparation of photo-masters in microelectronics, (4) manipulation of probes on integrated circuit chips, (5) manipulating micro-electrodes into nerves—in this application the rapid step action is expected to assist in entering the membrane, (6) microtome sectioning, (7) moving mirrors in laser beams, and (8) grating ruling in conjunction with a laser measuring device.

Using ceramic tube 1in long, type PZT-5A (made by Brush Clevite Co. Ltd.), a sensitivity of 500V per μ m can be obtained.

Electronic-fluidic interface switch

Electrical two-state signals from digital or switching systems can be converted very directly into corresponding air pressure signals by an unusual switching device shown by the University College of North Wales. It works on the principle that a laminar jet of air issuing from a tube can be made turbulent by a small continuous disturbance close to the tube. The disturbing mechanism in this case is the ion wind produced by a corona discharge occurring between a point and a plane electrode, and this forms the electrical input to the interface switch. The fluidic output



Electronic-fluidic interface switch, showing jet tube on left, discharge electrodes in the middle and collector tube on the right. (University College of North Wales).

signal is obtained by positioning a collector tube coaxially with the air jet so as to sense whether it is laminar or turbulent. The laminar jet diverges slowly and gives a large flow of air into the collector tube, whereas the turbulent jet diverges quickly and only a small flow passes into the collector. The corona discharge therefore switches the fluidic device from a high-output to a low-output state. Tests so far have shown that an electrical signal of 3kV passing a current of 0.1μ A will give a fluidic signal strong enough to operate commercial fluidic devices. In the demonstration a water gauge tube was used to indicate the pressure signal obtained. The device is said to be suitable for operation over a wide temperature range and in noisy and dirty environments.

Neon matrix display tubes

Glow discharge alpha-numeric display tubes based on the matrix principle were shown by Mullard Research Laboratories. Characters are formed by the selection of dots in a 7 \times 5 array, each dot being the negative glow at a cathode recessed below the glass surface of the tube. The cathodes are arranged in a thin, flat assembly while the common anode, a fine wire mesh, is between the cathodes and the viewing window. The tube is filled with neon gas. Cathodes can be selected to form characters by means of a diode decoding matrix, and, for example, to display the ten numerals a matrix of 40 diodes would be needed. Each neon dot element has a typical breakdown voltage of 170V, a maintaining voltage of 130V and a current consumption of about 150 μ A, so a character requiring, say, 16 dots out of the



Glow discharge alpha-numeric display tubes (Mullard Research Laboratories). The picture shows three of the tubes mounted on a printed wiring board.

available 35 would draw about 2.5 mA. The luminance of the display—which is more than adequate for nomal room use—is in excess of 2,000 cd/m². The tube gives a character size of $10 \text{ mm} \times 7 \text{ mm}$ and is made in a form which enables it to be mounted on a printed circuit board.

Camera to computer, direct link

Pattern recognition is becoming of increasing importance in a number of extremely diverse fields. One could almost say that every computer user would find it useful to be able to feed in data by optical means. E.M.I. showed the results of some work they have been doing in this field. A television camera was connected directly to an ICT 1905 computer via a standard interface. The programme was such as to sample the input picture at the 1,600 points in a 40 \times 40 matrix. The signal level, or picture brightness, was measured at each of these 1,600 points and assigned a value between one and 64, which was stored.

In the experiment a graph plotter was programmed to deposit ink in six different density levels. The six density levels were achieved by rather crudely adjusting the amount of shading in a given area.

The 64-level pattern in the computer store was reduced to six levels and fed to the graph plotter. In spite of the rather rough treatment the original signal from the camera had been subjected to, the result at the graph plotter was recognizable. Admittedly all detail had gone but the various areas of dark and light were clearly there. If the graph plotter could have accurately reproduced the six levels of shading the results would have been much better.

All this goes to show that a large amount of information can be rejected and yet the picture can still be recognized. The object of this work is to discover just how much detail machines will require in order to distinguish between different objects. When this has been decided—and the results will depend upon the particular application—we can expect to see special purpose pattern recognition machines on the market that do not require a full-size computer to drive them.

Seismic pattern recognition

A combination of threshold detectors and timing circuits is used in a perimeter alarm system (called AIDA), shown by Elliott's, for detecting human footfalls even when these are masked by other vibrations having similar frequency spectra.

Analogue signals from an array of geophones buried round the perimeter of the protected area are amplified and filtered to remove frequencies outside the spectrum of interest. The filtered signals are then applied to pattern recognition circuits, which measure rise- and fall-time, duration, amplitude and repetition rate, and if you prefer to monitor with an oscilloscope instead of a meter...

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compare the results with the normal characteristics of human footfalls stored in the equipment. Simultaneous agreement on all measurements for any particular channel will result in the alarm for the particular area being actuated.

Many uses for such alarm systems can be envisaged. For instance, the equipment could be made to recognize the normal vibration pattern of a machine tool and any departure from this pattern could be used to shut down the machine or sound an alarm—and the equipment could function along the clatter normally associated with a shop floor.

Full cycle thyristor firing

The conventional way of using a thyristor to control power in a load is to use phase control. With this method the firing of the thyristor is delayed after the beginning of the half cycle by an amount which depends upon the load power required. Firing late in the cycle gives low power, and early firing gives a high power. Now while current always flows at the end of a half cycle, it seldom does at the beginning, and this gives the effect of a lagging power factor. Also the supply current waveform is poor, containing many harmonics. These harmonics are produced into the MHz region and can cause severe radio interference.

These disadvantages can be overcome to some extent by the adoption of 'burst firing'. Here the thyristor is turned on at the beginning of a half cycle and remains on for several cycles. It turns off for several more cycles before turning on again. The average power is varied by varying the ratio of the number of 'on' cycles to the number of 'off' cycles.

Derby and District College of Technology have designed and built a logic system to control a triac (bi-directional thyristor). A five-bit binary number is the command signal, the triac turning on for as many cycles in a block of 32 as the binary number represents In general the conducting cycles are spaced reasonably evenly through the 32 cycle period. This is an advantage over normal burst firing methods. The need for a digital command signal may be a disadvantage in many systems but in the case of a computer controlled process a digital signal is produced so that a computer could directly control a heating element in a process control system.

Vibrating wire clinometer

A clinometer employing a rather interesting principle was shown by Smail, Sons and Co. Ltd. (agents for Firma H. Maihak, A.G.). A steel wire is stretched between an oil damped pendulum and a fixed point on the instrument's casing, so that movement of the pendulum with change of slope alters the mechanical tension on the steel wire.

The steel wire is the frequency determining component in an oscillator; any change in the mechanical tension alters the resonant frequency of the steel wire and, therefore, the oscillator frequency.

Any variation in the angle of the instrument's casing results in a change of oscillator frequency, which is measured at a remote point. Using this principle clinometers are available that can measure angle changes as small as 2 seconds of arc.

High sensitivity photoemitter

A photoemitter being developed by Mullard Research Laboratories is claimed to have a sensitivity greater than that of any existing photocathode. It is formed by exposing the surface of single-crystal p-type gallium arsenide to the gases caesium and oxygen. Under optimum conditions the photoemitter has a sensitivity in excess of 500μ A per lumen, and it has a response which extends through the visible spectrum and into the infrared as far as 0.9μ m. The example on show was one of these photoemitters combined with a channel electron multiplier to form a



Photomultiplier tube using a new high-sensitivity photoemitter. (Mullard Research Laboratories)

compact, highly sensitive photomultiplier. There is a possibility that the response can be extended farther into the infrared by the use of other chemical compounds.

H. F. Predictions-May

The prediction charts show median standard frequency (MUF), optimum traffic frequency (FOT) and the lowest usable frequency (LUF) for reception in the U.K. Unlike MUF, the LUF is closely dependent upon such factors as transmitter power, aerial gain and type of service. LUFs shown were drawn by Cable and Wireless Ltd, for commercial telegraphy, using several kilowatts and aerials of the rhombic type.

Seasonal changes are most striking on the Hong Kong route; the peaks of recent months are depressed giving an FOT below 20MHz which changes little throughout the 24 hours. The Montreal route shows the same characteristic as it is also an East/West path in the same hemisphere. Daylight FOTs for the trans-equatorial paths continue above 20 MHz.



New Television Camera and Recorder Techniques

Highlights from the National Association of Broadcasters' convention in Washington, U.S.A.

by Aubrey Harris,* M.I.E.E.

The forty-seventh annual convention of the National Association of Broadcasters was held from March 23rd to 26th in Washington, D.C., U.S.A. The convention consists of a large number of equipment exhibits (146 booths) and many programme and engineering meeting sessions. The quality of the technical papers at N.A.B. is seldom worthy of note, whereas the broadcast equipment manufacturers invariably burn their midnight oil in the weeks before the convention in order to be able to show their latest electronic endeavours to the assembled, expectant engineers and station staff. Incidentally, to give an idea of the size of the industry in the U.S., the F.C.C. has just announced that in 1968 there were 4,203 a.m. stations, 2,198 f.m. and 811 television stations (excluding translators).]

The 1969 show ran true to form, the main highlights being in innovations in colour cameras and videotape recording equipment.

Cameras

There were two major introductions in the colour camera field, one at the high-cost end of the scale and one in the low-cost region. The new Philips PC-100 three-tube (lead oxide) camera at \$84,850 (£35,000) has several striking new features, the most astounding of which is that the camera cable consists of one single triaxial cable (that is, a coaxial cable with two separate, insulated screens). This single cable, about $\frac{3}{16}$ in diameter, carries not only the encoded N.T.S.C. output signal and monitor signals from the camera to the camera control unit, but also sixty digital and sixty analogue functions, external viewfinder control signals, intercommunication, programme sound and camera power (100 volts d.c.) from the c.c.u. to the camera itself.

This not inconsiderable feat is accomplished by three separate multiplexing channels on the cable: a maximum length of 1 mile can be used normally, although this can be extended by the use of repeater amplifiers. The weight of the cable is only about onetenth of that of the conventional colour camera cable, and this together with its smaller size is bound to have significant advantages in the logistics and costs of outside broadcast operations and studio design.

The camera features a colour bar generator, an encoder and a contour enhancer, all built into the camera proper. The camera head, without lens, weighs only 70 lb and measures a mere $19\frac{1}{2}$ in. long, 17 in. wide and is $15\frac{1}{2}$ in. high (to the top of the viewfinder)—remarkably compact for a three-tube camera. The viewfinder tilts up and down and has quite a novel feature. It may be turned in a horizontal plane through 90° so that the picture image may be viewed normally from the side of the camera. Those who have attempted to make adjustments on the side of a camera while viewing a picture at the rear will appreciate the advantage of this.

Registration accuracy of 0.2% is claimed for the centre of the scanned image and accuracy of picture geometry within 0.5%of picture height. The peak signal to r.m.s. noise ratio is 50dB.

At the other end of the scale, RCA showed the PK-730, a new single-tube colour camera costing, without viewfinder but with 6:1 zoom lens, 6,500 (£2,600) and with electronic viewfinder 9,850 (£4,080). The camera utilizes an 8507A vidicon, in the optical path of which there is a pair of striped colour filters; the spatially shared signals representing the colour analysis of the viewed scan are electronically processed and encoded into an "N.T.S.C.-type" signal for helical scan recording or closed-circuit transmission. The encoded output signal differs from the true N.T.S.C. waveform in that the sub-carrier is not locked to the line frequency, and instead of I and Q signals of different bandwidth being used to modulate the sub-carrier, R-Y and B-Ysignals of identical bandwidth are used for the quandrature/colour modulation. This simplified arrangement avoids the need for a delay line and compensating circuitry, although it was not claimed that the signal was up to full broadcast standard; nevertheless the subjective result was very acceptable and this type of system seems bound to find acceptance in many educational, commercial and other less critical uses.

Incidentally, the output can be made to conform to N.T.S.C. (and presumably PAL) standards by the addition of the appropriate encoder.

Video recording

A new entry into the professional videotape market was Westel who showed a new oneinch helical scan recording system in yet another tape format. This recorder uses a small diameter (2-3 in.) drum guide with an "omega wrap" (200°) two-head configuration. It differs from other helical scan recorders in that only one-sixth of the information in a field is recorded per head sweep (as compared to complete fields in most other helical scan machines and compared to between 10 and 17 lines in transverse scan, quadrature-head type videotape recorders). Either longitudinally oriented or laterally oriented videotape, it is claimed, may be used with little, if any, difference in performance.

Another innovation is a dual-capstan tape device maintaining constant tape tension into, within, and out of the video head recording path. One problem with helical-scan recorders has always been the difficulty of maintaining correct and consistent tape tension around the drum guide; the dualcapstan arrangement, which has previously been used on computer-type, instrumentation and some audio recorders, simplifies the problem of consistent tape tension without external air lubrication. The reel-to-reel tape speed is 15 in. per second and a head-to-tape writing speed of 1260 in. per second is used. A total of five printed-circuit motors is used on the rack-mounted transport, one each for the supply and take-up reels, one for the head drum and one for each of the two capstan motors.

The time-stability of the video signal, directly from the demodulator, is in the region of ± 250 nanoseconds. This brings it well into the capture range of electronic timebase correction devices (such as Amtec) which are used both for "picture-straightening" purposes and also for bringing the signal within the range of colour time-element compensators (such as Colortec) for direct colour signal recovery operation.

Full interchangeability is claimed between the two versions available—the "recordonly" WR-250, weighing 37 pounds including batteries, and the rack mounted studio unit WRR-350.

The output video signal-to-noise ratio is said to be 5 to 6dB higher than in other existing studio quality videotape recorders, although no exact figures were quoted. The U.S. price of the studio colour recorder (including the necessary time-element and colour compensation circuitry) is \$58,000. (\pounds 24,000); the monochrome version price is \$33,500 (\pounds 13,900). The portable, "recordonly," versions sell for \$19,500 (\pounds 8,300) and

University of California, U.S.A.

Wireless World, May 1969

\$17,500 (£7,350), colour and monochrome respectively.

These recorders combine the simplicity of the helical-scan type of recorders with the stability of the broadcast type machine into a low-cost high-quality v.t.r. However, it seems unlikely that broadcasters would wish to be involved in yet another television tape standard, with all its attendant problems, particularly with the likelihood of broadcast E.V.R. (Electronic Video Recording) being introduced in a year or two.

The final step enabling absolute synchronization of two or more videotape recorders seems to have been reached by Ampex, who showed for the first time their RA-4000 Random Access Programmer. This enables tape "addresses" on different portions of different tapes to be selected remotely, automatically cued-up and the machines run sychronously from that point, enabling precise editing to be done by programmed information put into the RA-4000 or by treating the two playback machines as inputs to a video mixer the output of which feeds a v.t.r. in the "record" mode.

To enable this to be accomplished the tapes to be operated on have recorded on the second audio (cue) track digital addresses which indicate hours, minutes, seconds and also individual frames. The address recording may be done prior to, during or after the programme video and audio tracks are recorded. During operation the addresses at which the two machines are required to be run in frame sync are keyed into a control panel, as is also the address of the frame at which the edit from tape A to tape B is to be made.

The machines are started and an automatic search is made for the addresses keyed in, indicating the start of the sections required to be run in picture synchronism. The recorders find these addresses and then backup to a cue point 150 in. (10 seconds) in advance of these points; the recorder reaching its cue point first waits for the other to cue up. Then both advance at nominally playing speed while the two addresses are continuously compared; any discrepancy is detected and corrected by automatic operation of the capstan tape-speed override circuit on the logging machine. The machines are then in picture synchronism at the desired frame; the tape speed control then reverts to normal intersync operation.

Another departure in the realm of videotape recording, which was likened to a "v.t.r. juke-box", was the RCA colour videotape cartridge recorder/player; this has been designed to record and replay short sequences (up to three minutes long) of video and audio material. The machine can accommodate eighteen enclosed cartridges of 2-in wide videotape, which may be recorded and reproduced by the two tape transport mechanisms within the equipment. Cartridges are reproduced under command of either external signals or signals recorded on the second audio track of a cartridge being played. While one cartridge is being reproduced by one transport the next cartridge is being automatically cued-up ready to play.

The switching signal transfers the signal system input connection from the transport playing the first cartridge to that handling the second cartridge; this gives a virtually instantaneous picture switch and is ideal for running back-to-back commercial sequences, or other short programme sequences.

Automatic programme control

Both of the two last-mentioned devices seem to lend themselves admirably to incorporation into equipment for automatically controlling the sequence of television programme material being fed to the transmitter or network. Many companies were showing such equipment-General Electric, Ampex, Central Dynamics, RCA, Sarkes-Tarzian, Visual Electronics, to name a few. Details of each company's system vary, but in general these automation systems allow a large number of programme segments to be run, in rapid sequence if required, without manual action by an operator.

For example, film projectors, videotape and audio recorders may be cued-up and started at precise times, slides may be changed, and fading, mixing and cutting between video and audio sources is carried out automatically, although provision is always made for manual override, in case of unforeseen circumstances or because of a change of programme scheduling.

In a typical system each operation is keypunched into an IBM punched card, indicating video source, audio source, type of transition, time of transition, segment duration and brief title description. The cards are then stacked and fed into a card reader which feeds the information in sequence to a control unit, which operates in conjunction with the station clock system. Very often the "next ten" operations are displayed by a character generated display on a picture monitor, enabling the supervisor to check the forthcoming events and change them when desired. Video tape recorder and projector "pre-roll" cues are automatically given to enable these sources to be stabilized by the time the control equipment is ready to switch them to transmission.

Some more sophisticated systems take their input from the station computer, disc file, magnetic tape or remote lines rather than directly from a card reader. A local teleprinter type keyboard is often adjacent to the control equipment for insertion of special commands.

Conferences and Exhibitions

Further details are obtainable from the addresses in parentheses

LONDON

- May 6-8 Savoy Place Power Thyristors and their Applications
- (I.E.E., Savoy Pl., London W.C.2) May 20-23 Electronic Component Show (Industrial Exhibitions, 9 Argyll St., London W.1)
 - www.americanradiohistory.com

May 20-23	Kensington Close Hotel, W.8					
Electronics	Exhibition					
(T. Jeffrey I	Burton Associates, 198 Forest					
Road, Tun	bridge Wells, Kent)					
May 28-29	Northern Polytechnic					
Computer	Aided Design Techniques					
for Electronic Circuits						
(Dept. of Ele	ectronic and Communications					
Eng, North	ern Polytechnic, Holloway,					

London N.7) BIRMINGHAM May 2 & 3 Grand Hotel Service—its place in Marketing

(Society of Service Managers, 1 Tichborne Close, Frimley, Surrey)

EASTBOURNE

May 6 & 7 Grand Hotel Automated Inspection (Scientific Instrument Research Assoc.,

South Hill, Chislehurst, Kent BR7 5EH) OVERSEAS

May 5-7

Electrical & Electronic Measurement W. J. Moore, 797 Dunloe St., Ottawa 7) May 5-7 Dallas

Microwave Symposium (I.E.E.E., 345 E.47th St., New York, N.Y.10017)

May 5-8 Farmingdale Instrumentation in Aerospace Simulation

(I.E.E.E., 345 E.47th St., New York, N.Y.10017)

May 6-8 Ispra Nuclear Electronics (Prof. Luciano Stanchi, C.C.R. Euratom,

21020 Ispra, Italy) May 6-8 Atlantic City

Frequency Control

(M. F. Timm, Electronic Components Lab., U.S. Army Electronics Command, Fort Monmouth, New Jersey 07703)

May 7-9 Artificial Intelligence

(British Computer Soc., 23 Dorset Sq., London N.W.1)

May 14-28 Moscow

Automation '69

(Scientific Inst. Mftrs' Assoc., 20 Peel St., London W.8)

May 19-21 Dayton Aerospace Electronics Conference

(I.E.E.E., 345 E.47th St., New York,

N.Y.10017) May 19-23 Montreux

Television Symposium & Exhibition (Secretariat, Case-Box 97, 1820 Montreux)

May 21-23 Edmonton Microwave Power Symposium

(W. R. Tinga, Elect. Eng. Dept., University of Alberta, Edmonton, Alta)

May 21-23 Gaithersburg Electron, Ion, and Laser Beam Technology

(Dr. L. Marton, National Bureau of Standards, Washington, D.C. 20234) May 22-23 Washington

May 22-23 Applied Magnetics (I.E.E.E., 345 E.47th St., New York; N.Y.10017)

May 26-28 Washington Laser Engineering and Applications (Lewis Winner, 152 W.42nd St., New York; N.Y.10036)

Ottawa

London Component Show

Provisional list of exhibitors at the international show in May

The biennial Electronic Component Show opening at Olympia, London, on May 20th for four days is the 21st in the series sponsored by the Radio & Electronic Component Manufacturers' Federation. It will be the biggest of the series and will be the first international show sponsored by the Federation.

We list below the manufacturers and agents who have taken space. Indented below the names of U.K. agents are the overseas companies whose products they will be exhibiting. Wireless World will again be exhibiting and on our stand we will be demonstrating the Logic Display Aid to be described in the series of articles which begins in this issue. It is planned to include in our July issue a selection of the new components, instruments and materials introduced at the show. On the opposite page are illustrated a few of the new products already announced by exhibitors.

During the last two days of the exhibition a semiconductor symposium is being held in the Pillar Hall at Olympia under the auspices of VASCA, the Electronic Valve and Semi-Conductor Manufacturers' Association. The morning session on the 22nd will be concerned with linear integrated circuits (from basic economics to the latest developments), the afternoon session with digital i.cs; and the morning session on the 23rd with power devices. Admission to each session is by ticket, costing £1 per session. Details of the programme and also tickets are available from VASCA, Mappin House, 4 Winsley Street, London W1N ODT.

The show opens from 10.00 to 18.00 daily and admission costs 5s.

AB Electronic Components **AEI** Semiconductors A.K. Fans AMP Industrial Air Control Installations Aladdin Components Aladdin Electronics Alma Components Alston Capacitors Amphenol Ancillary Developments Ariel Pressings Arrow Electric Switches Ashburton Resistance Co. Astralux Dynamics Autronic Developments Avel Products Aveley Electric Avo/Taylor

BICC-Bumdy B. & R. Relays Adams & Westlake Co. (U.S.A.) Benedict & Jager (Australia) Gordos Corp. (U.S.A.) Minimotor S.A. (Swltzerland) Siemens A.G. (W. Germany) Versa N.V. (Holland) BSR **Bakelite Xylonite** Barlow-Whitney Beckman Instruments Bedco Bedco Belclere Company Belling & Lee Berec International Bird Electronic Bonnella, D. H., & Son Bowmar Instruments Bradley, G. & E. Bradauer, C., & Co. Uniform Tubes Inc. (U.S.A.) Brit. Insulated Callender's Cables British Physical Labs. Brookdeal Electronics Brown, A. G., Electronics Brush Beryllium Co. Bulgin, A. F. & Co. Burgess Micro Switch Co

C.C.L. C.G.S. Resistance Co. C.I. Automation Cadmium Nickel Batteries Cambion Electronic Products Cannon Electric (G.B.) Carr Fastener Cathodeon Cathodeon Crystals Centralab Chance-Pilkington Channel Electrical Equipment Air-LB Ciba (A.R.L.) Circuit Integration Circuitape Clare, C. P. Clare-Elliott Clarke, H., & Co. Coil Winding Equipment Co. **Cole Electronics** Aumann. W. (W. Germany) Kumag A.G. (Switzerland) Slemens A.G. (W. Germany) Colvern Computer Controls Computing Techniques Concordia Electric Wire Connollys (Blackley) Corner, G., & Co. Cosmocord Counting Instruments Crouzet England cie Petercem (France) Schmersal, K. A., & Co. (W. Germany) Culton Instruments Darby Industries Data Precision (Equipment). Davall, S., & Sons Davu Wire & Cables Daystrom Deac (Great Britain) Dial Engineering Co.

Diamond H Controls Digital Equipment Corp. (UK) Dubilier Condenser Co Dudleys (Redditch) Du Pont de Nemours International

EMI Electronics EMI Sound Products East Grinstead Electronic Components Eddystone Radio Egen Electric Ekco Plastics Elcom Electrautom Electro Acoustic Industries Electrolube Electro Mechanisms Bytrex Inc. (U.S.A.) Kulite Semi-Conductor Prods (U.S.A.) SFIM (France) Sakae Tsushin Kogyo Co. (Japan) Schaevitz Engg. (U.S.A.) Tokyo Sokki Kenkyujo Co. (Japan) Electrographic Electroprints Electrosil Electrothermal Engineering Electroustic Firma Frako (W. Germany) Hirschmann (W. Germany) W. Ruf Ohg (W. Germany) Elliott-Automation **Elliott Brothers** Enalon Plastics Enfield Phelps Dodge Engineering Enterprises English Electric Valve Co. Enthoven Solders Erg Industrial Corp. Erie Electronics Erma Ether Ever Ready Company Evershèd & Vignoles Dynamic Insts. Corp. (U.S.A.) Gossen, P., & Co. (W. Germany) S.F.A.I.R.E. (France) Tettea A.G. (Switzerland)

Dymar Electronics

Fane Acoustics

Farnell Instruments Ferranti Filhol, J. P. Fine Wires Flight Refuelling Floform Parts Formica Formica Fothergill & Harvey French composite display

G.I. Microelectronics G.K.N. Screws & Fasteners Gardners Transformers General Instrument Group General Instruments Girdlestone Electronics Goodmans Loudspeakers Greca Products Greenpar Engineering Guest Electronics

Haddon. Thomas & Stokes Haddon Transformers Hallam, Sleigh & Cheston Harrison, A. T., & Co. Harwin Engineers Hawthorn Baker Healey Meters Heberlein & Co. Hellermann Hellermann Deutsch Hengstler, J., & Co. Henry & Thomas Hesto (Henkels-Stocko) Hewlett-Packard Heyco Manufacturing Co. Highland Electronics Hilger Electronics Hilger & Watts Hinchley Engineering Co. Hivac Honeywell Controls Hopt Electronics Howells Radio Huber, J. J. Hysol Sterling

Imhof, Alfred Imperial Chemical Industries Insulating Components & Materials Instn. of Electrical Engrs

Jackson Brothers J. Beam Engineering Jermyn Industries Jidenco Joseph Electronics Duerwaechter-Doduco (W. Germany) Electrovac (Austria)

Keithley Instruments Keyswitch Relays Klippon Electricals Knowles Electronics Kolectric Coil Winding Equip. Co. (U.S.A.) Midland Eng. & Manf. (U.S.A.) R.M.T (Italy)

L.C.R. Components Lectropon Amelco Semiconductors (U.S.A.) Diodes Inc. (U.S.A.) Globe Industries Inc. (U.S.A.) I.E.R.C. (U.S.A.) Sage Electronics Corp. (U.S.A.) Soriau et Cie (France) Lee Green Precision Industries Levell Electronics Linton & Hirst Litton Precision Products Londex London Electrical Mfg. Co. Lucas, Joseph (Electrical) Lustraphone Lyons. Claude ABEM Inst. Co. (Sweden) Bishop Inst. (U.S.A.) Control Data Corp. (U.S.A.) Electrons Co. (U.S.A.) Elgenco Inc. (U.S.A.) Guildline (nsts. (Canada) Hallmark Standards Inc. (U.S.A.) International Light Inc. (U.S.A.) Millivac Insts. Inc. (U.S.A.) Rockland Labs. Inc. (U.S.A.) Straumann, R. (Switzerland) T.R.G. Inc. (U.S.A.) Lyons Instruments

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M.B. Metals M.C.P. Electronics McMurdo Instruments Magnetic Devices Magnetic & Electrical Alloys Mallory Batteries Mann Components Mansol (Great Britain) Marconi Company May Precision Components Metway Electrical Industries Micro Waves Inst Microwave Associates Midland Silicones Milton Ross Co. Mitsubishi Electric Corp. Morganite Resistors Motorola Semiconductors Mullard Muller, Dr. Kurt Multicore Solders Murex

N.S.F. Newmarket Transistors Newport Instruments

O & W Electronics Oliver Pell Control Oltronix U.K. Ospec Oxley Developments Co.

Painton & Co. Palmer Aero Products Palmer, G. A. Stanley Arco S.p.A. (Italy) Collins Radio Co. (U.S.A.) Republic Electronics Corp. (U.S.A.) Resista GmbH (W. Germany) T.E.C. (France) Park Royal Porcelain Parmeko Pedoka Perivale Controls Co. Permanoid Permark Philbrick/Nexus Research Planer. G.V Plannair Plasmoulds Plastronics Plessey Company Plex (Engineering) Precious Metal Depositors Helmut Fischer (W. Germany) P.M.D. Continentale (France) Schlottar, Max (W. Germany) Precision Electronic Terminations Pressac Pye of Cambridge Pye Switches

Quickdraw Co.

Radiall Microwaye Components Radiatron Rank Precision Industries Rathdown Industries Raychem Redpoint Reliance Controls Rendar Instruments Research Instruments Resistances **Rivlin Instruments** Rola Celestion Rosenthal Technical Components Ross. Courtney & Co. Royal Worcester Ind. Ceramics

SASCO SGS (U.K.) STC/ITT Components Group STC Semiconductors S.T.P. Electronics Salford Electrical Instruments Sarcem Products Satellite Engineering Schjeldahl Co. Sellotape Products Service Electric Co. Shure Electronics Sifam Electrical Instrument Co. Signetics International Corp. Simmonds Relays Sintered Glass to Metal Seal Co.

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- Smart & Brown Connectors Smiths Industries S. London Elec. Equip. Co. Southern Transformer Products Spear Engineering Co. Stability Capacitors Stadium Standard Telephones & Cables Steatite Insulations Steatite Insulations Steatite & Porcelain Products Suflex Suhner Electronics Surry Steel Components
- Technograph & Telegraph Technology, Ministry of Tectonic (Electronics) Tektronix U.K. Telcon-Magnetic Cores Telcon Metals Telephone Manufacturing Co. Telequipment Telford Products Termco Terminal Insulators Thorm-AEl Radio Valves & Tubes Thorn Bendix

Thorn Electrical Industries 3M Company Transformer Windings Transitron Electronic Tucker Eyelet Co.

Ultra Electronics (Components) Union Carbide UK

Valory Watch Co. Varelco Varian Associates Veeco Instruments Venner Electronics Vero Electronics Vision Engineering

Watson, W., & Sons Waycom Wego Condenser Co. Weir Electronics Weller Electric Welwyn Electric West Hyde Developments Westinghouse Brake & Sigral Co. Weyrad (Electronics) Whiteley Electrical Radio Co. Wingrove & Rogers Wire Products & Machine Design Woden Transformer Co: Wolsey Electronics

Z & I Aero Services Zenith Electric Co.

Typical of the series 40 stabilized power supplies introduced by A.P.T. Electronic Industries, is this 43D10 which provides 50V at 10A. WW 393 for further details





Lyons Instruments new high-power pulse generator (PG 25) provides two outputs of opposite polarity each capable of supplying 1A into $SO \Omega$. Frequency range is 10 Hz to 20 MHz.

WW 391 for further details

The secondary bobbin in this small mains transformer (Hinchley Engineering) incorporates a moulded skirt which shrouds the primary winding. WW 394 for further details





The infinite resolution contactless potentiometer recently introduced by Salford Electrical Instruments is available in various configurations to measure either angular or linear displacements.

WW 392 for further details

This d.c. multimeter from Levell Electronics measures $0.3 \ \mu V$ to 1 kV in 18 ranges, from 1pA to 1A in 24 ranges and from 0.3Ω to 1000 M Ω in 18 ranges. It incorporates a high-gain, solid-state, chopperstabilized d.c. amplifier. WW 396 for further details





Low torque, rotary-action micro switches (type VII) introduced by Burgess Micro Switch Company for sensing, detection, counting and similar applications where precision coupled with resistance to nechanism derangement is essential. WW 395 for further details

Circuit Ideas

Transistor a.c. mains controller

The circuit shown in Fig. 1 is self-explanatory. The output waveform is approximately sinusoidal. Output power is, of course, dependent on the two transformers and on the allowable dissipation of the transistors. Care will need to be exercised in operating the ganged 1k Ω bias control. As a precaution a suitable

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'stopper' resistor can be put in series with each of the two variable resistors to limit maximum dissipation.

J. R. HARRIS, Woodstock, Oxon.

J

n

15405 or

similar

Constant-impedance attenuator

This circuit presents a relatively constant impedance in one direction. It was originally devised as a simple volume control for use with a loudspeaker, as shown in Fig. 1. With VR₁ equal to 3R the input impedance varies between 75% R and 120% R. With VR equal to 4R it varies between 80% R and 133% R. This circuit can be used, for example, to improve the attenuator of the "Low Distortion RC Oscillator" by P. F. Ridler (Wireless World, August 1967). The attenuator originally used (Fig. 2) will only give an output at 600Ω in the 2.2V position if the pot wiper is in the middle of the track, the impedance falling to 350Ω at either end. These variations can be much reduced by replacing the attenuator with the circuit of Fig. 3.

D. AUSTIN, ATV Network Ltd., Birmingham.



Fig. 1. Constant impedance attenuator.

350

An a.c. mains power

Oscillator using operational amplifier

RV1b

OC36

similar

simple Wien bridge audio oscillator A can be built using an integrated circuit operational amplifier. The bridge is connected between the output and the noninverting input of the op. amp. A thermistor and resistor connected between the output and the inverting input limit the output amplitude, producing a sinusoidal output. Two additional capacitors and one resistor are necessary to control the high-frequency gain of the op. amp.



The circuit shown oscillates at about 1kHz with an output of 3.5 volts p-p into a 100 ohm load. The output resistance is very low. Operation from other power supplies or a single supply is possible. The normal audio frequency range can be covered by the usual arrangement of switched capacitors and a two gang potentiometer. D. W. J. BLY, Mullard Observatory,

Cambridge.





Fig. 2. Attenuator originally used in Ridler's oscillator.



Fig. 3. Improved attenuator for oscillator.

Marconi Radio Telephone Terminal Type H5510

For interconnecting h.f radio circuits with inland telephone networks.

- Eliminates circuit loop instability caused by feedback from the receive-to-transmit path.
- Reduces the effects of fading and high noise level, characteristic of h.f radio propagation.
- Controls outgoing speech so that transmitter loading is maintained at optimum level.
- Channel displacement and privacy equipment can be incorporated.
- Maximum capacity, 4 channels.

Solid-state circuitry Plug-in modulator book units Flexible design Optional built-in tape recorder



Marconi telecommunications systems

Member of GEC-Marconl Electronics Limited

The Marconi Company Limited, Radio Communications Division, Chelmsford, Essex, England

Better Trend-Test your Data Transmission Systems



The original Trend data transmission test set type 1, was the best instrument in its field: everybody – including the GPO liked it. But we've improved it.

The new data transmission test set type 1-3 has many improvements, including a variable error threshold control and an extra counter to count pseudo random blocks in error or the total number of blocks received or to extend the error count to 2047x10⁶. Trend dont just keep up with the field, they lead it.



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www.americanradiohistory.com

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The Editor does not necessarily endorse opinions expressed by his correspondents

Resistance labelling

When resistor values are marked on circuit diagrams in the usual abbreviated way; e.g., 6.8k for 6,800 ohms, errors can arise from the omission of the decimal point. (This can very easily happen if the diagram is a poorly duplicated photocopy.)

The risk of this type of error can be avoided by using the Continental convention whereby the 'multiplier' letter is substituted for the decimal point. Thus, 6.8k becomes 6k8, 3.3M becomes 3M3, and so on. This notation seems to be catching on among British engineers whose work brings them into contact with Continental circuit diagrams, and this encourages me to suggest that the time may be ripe for journals such as Wireless World to adopt it.

Naturally, it would look a bit odd at first. Any new notation does. But we don't bat an eyelid at 'kHz' these days, which shows how quickly one gets used to things.

The system appears to break down when one has to indicate fractions of an ohm, as in, say, 3.9 ohms. One could put 'U' for 'units' and write 3U9, but the existing Continental practice seems to be to write '3E9', etc. I don't know what the 'E' stands for (I would guess 'Einheit' though perhaps some better-informed reader can supply the answer) but one could still use the system even it meant nothing in English.

G. W. SHORT,

South Croydon.

Aerial erection

At 40 miles from London I've always found a home-made aerial in the loft quite satisfactory for TV and f.m.-mono transmissions. But now having assembled a stereo version I find I need an outside aerial, and for me at any rate that means a factory built job. I took it for granted that dealers would carry them as a matter of course: they always used to. But not so now. I've been to half a dozen dealers in three sizeable towns and though some are a bit cagey on the subject it appears that (a) aerials are practically not available retail, and (b) there is virtually only one aerial erecting contractor for the greater part of Essex and he is at the other end of the county, which would inevitably put the cost of an erected aerial sky-high.

I am writing to you as Editor of our senior radio magazine because I feel sure your important advertising links with the trade will not prevent you ventilating a subject of some importance to the majority of your readers. No doubt for dealers it is a great convenience to be able to put this work out to contract, but if as a reciprocal gesture they undertake to refrain from selling aerials retail, it seems to be a gesture at the public expense which the Monopolies Commission should take an interest in. But I shall be much interested to hear your views. P. I. A. INNES.

Dunmow, Essex.

Class A versus class B

I was very interested to read J. L. Linsley Hood's remarks concerning subjective differences between class A and class B solidstate amplifiers ("Simple Class A Amplifier", April) since I, too, have recently been performing a series of tests between the two classes of operation.

I feel that the 'slight edginess' referred to in the article is the subjective effect of crossover distortion which, in quite a few of the lower price class B amplifiers, tends to rise in magnitude with reducing power yields and with increasing frequency. High string notes whose harmonics run up to quite high orders, therefore, would be more affected by this sort of distortion than lower frequency signals of greater power (sound intensity). This, indeed, is one of the worst of class B subjective effects and by studying the harmonic yields over the frequency spectrum one would conclude that the crossover effect produces a whole series of oddnumbered harmonics extending to high orders which, of course, are singularly inharmonious to say the least!

The effect is aggravated by the extensive power bandwidth of many recent designs, especially those employing silicon transistors, power not uncommonly being delivered well into the 'radio spectrum' to 40kHz or more. While a passband in excess of the audible spectrum is desirable for maintaining waveshape and endowing good rise-time features, I feel personally that this power response business is being taken a bit too far. I like to roll-off at a fairly slow rate of 6dB/octave from about 25kHz, and by doing this (not necessarily with a switchable low-pass filter) the rise-time performance is not unduly upset (from the audio aspect) yet disturbing harmonics are deleted. It seems that although such harmonics might fall above the audible spectrum they can certainly interfere with the complex waveforms of music signals. Rolling-off in that way often minimises the 'listener fatigue', also mentioned by Mr. Linsley Hood, when the amplifier happens to be prone to such symptoms.

It is noteworthy that some listeners are less disturbed by class B amplifiers than others. But those whose ears do 'twitch' to slight high-order odd-numbered harmonic distortion will most certainly discern a difference when rapid A-B testing is made against a well designed class A amplifier and when the comparative material itself carries a high range of overtones and when the listening is performed at relatively low power level. It must be mentioned, though, that there are some class B designs-tailored to reduce crossover distortion-that sound virtually the same as a good amplifier in class A mode; but the price is an important factor that must be taken into account when making comparisons like this. From the consumer's point of view a low-cost class A amplifier-assuming that it can produce sufficient audio power and the other audio parameters are as required-must surely be better than a relatively more expensive class B amplifier, the expense of which is the result of the extra design and researches applied to achieve a class A performance, anyway!

GORDON J. KING, Brixham, Devon.

Improper oscillations in transistors

Although not a regular reader of your magazine and although I did not see Mr Pitt's original letter on 'improper oscillations in transistors' in the January issue, I am bound to comment on a statement in the reply to this letter by Mr. Vanderkooy in March.

Mr. Vanderkooy asserts that one resistor and one transistor can never cause a transistor to oscillate. It is the experience of the nuclear physics group here that this is far from true. For the past eight years we have been exploiting the fast switching behaviour of transistors when used in the avalanche mode and although there are only a few (2N914 is the best) which can be used reliably as stable pulse amplitude discriminators, we have found that almost all n-p-n transistors can be made to avalanche repeatedly without destroying themselves. These transistors only require a collector load sufficient to limit their dissipation and as the supply voltage is increased beyond the avalanche potential, the transistor bursts into relaxation oscillations analogous to the thyratron. In order to use the device as described the base has to be reverse biased to stop this oscillation and hold the transistor off. When an input pulse applied to the base exceeds the triggering threshold a collector pulse of 15V into a 50 Ω load with 0.5 ns risetime is common with the 2N914.

With regard to decoupling, it is very difficult to completely isolate the device as the radiation from perhaps a 0.5-A pulse rising in less than 1 ns is not easy to contain, and sealed boxes and elaborate supply filtering are required. Needless to say this transmission can be picked up easily on a small transistor radio.

As for capacitive loading of the collector, this decreases the pulse repetition frequency up to a point and then quenches oscillation.

Unfortunately the means to look at such phenomena are expensive, so might I suggest that one of your regular contributors designs a simple sampling unit to push the use of say a 10 MHz antique scope up to a few hundred MHz. This would not only open up a new domain of amateur investigation but no doubt please the 'Hams' of the 2-metre band.

N. W. BENNÉE, Dept. of Physics, Birkbeck College, London W.C.1.

Surface temperature ther-

With reference to the temperature measuring instrument described by Mr. L. Nelson-Jones in the April issue, there is a possible source of error which has not been mentioned.

This is when the instrument is used on an item the mass of which, over the contact area, is less than the conducting mass of the probe. In these circumstances the probe will conduct sufficient heat away from the contact area to lower the temperature of the contact area and give a low indication.

It is appreciated that Mr. Nelson-Jones has reduced the conducting mass of his probe to a minimum and that when used, as illustrated, on a large heat sink, errors from this source are negligible. However, if this probe were to be used to measure the temperature of, say, a 0.012in metal skin, the error may well be significant.

As this is so, I feel that this ought to be pointed out so that constructors may be warned of this limitation to accuracy. H. D. READ, Yeovil,

Somerset.

Jonneroett

The author replies:

I agree that with very small or very thin bodies a surface temperature measurement made with my instrument will be in error. This type of error is common to all such instruments, and the magnitude of the error depends on the relative heat losses of the probe and the body being measured in the case of small bodies. In the case of very thin bodies the error is due to the thermal resistance of the source in supplying heat to the heat losses of the probe. In my design I did try to keep the heat losses to a minimum for this type and size of probe because of this source of error.

I should perhaps have emphasised this aspect more strongly, and the fact that I designed the probe for the larger heat sink rather than for say individual TO5 transistors. The reason for this is that it is normal, for the smaller transistor, to measure the prevailing ambient temperature in the vicinity of the transistor, and then to calculate from the manufacturer's derating curves whether or not the transistor is in a safe operating temperature region. Whereas for larger heat sinks, due to the many factors involved, it is not as safe to calculate the temperature of the junction, mainly because one does not know the temperature of the heat sink on which the manufacturer's derating curves are usually based.

A useful but not infallible guide is the response time of the probe. If this is fast then the reading should be accurate, but if the pointer is very sluggish to take up a steady reading, then the reading should be regarded with suspicion.

L. NELSON-JONES.

Improving old loudspeakers

Mr. Bennet-Clark's novel theory (March issue) regarding the 'Hookean spring' action of loudspeakers seems strangely at odds with the rest of his text, in the course of which he advocates removal of as much spring as possible from the cone-edges of old loudspeakers.

All loudspeakers have a certain amount of axial restoring-force (spring) in their coneedge and coil-centring material, and this, whether corrugated paper or not, must be substantially 'Hookean' (strain proportional to stress, up to the elastic limit), if it is to last for any reasonable time.

One feels it is not failure of Hooke's Law which is to blame here, but rather that an oscillating mass attached to a spring has a natural period of vibration, the frequency of which is mainly determined by the mass /spring characteristics of the complete system. The stronger the spring in relation to the mass, the higher the natural frequency, and vice versa.

It follows therefore, that increasing the compliance, i.e., weakening the spring, and /or increasing the cone weight by giving it an extra skin, will both result in the lower fundamental resonance that Mr. Bennet-Clark has found in practice.

T. H. FRANCIS, High Wycombe, Bucks.

The author replies:

I quite agree with Mr. Francis' last paragraph and indeed, although I may not have expressed it clearly I appreciate that this was the effect of my treatment to old loudspeakers. The point that I wished to make

Our Next Issue

Amplifier survey. The terms and figures that contribute to the specification sheet of an audio amplifier will be thoroughly investigated by a design engineer accomplished in this field. Tabulated data on commercially available amplifiers will also be presented.

Units converter. As announced on p.202 a Wireless World designed "slide rule" for conversions between common electronics and radio units—frequency/wavelength, ratios/decibels etc. (including Imperial/metric SI conversions) will be offered to readers. about the use of corrugated paper surrounds was that they have elastic properties that are substantially non-linear; that strain is not proportional to stress.

I think that it is fairly generally accepted that either the use of cloth, leather, plastic foam or plastic roll surrounds substantially increases the linearity of movement of a loudspeaker cone. My aim has merely been to achieve this with an existing loudspeaker.

I do not feel, in answer to Mr. Francis' first paragraph, that I have said anything that constitutes a novel theory; all this seems to be accepted practice.

H. C. BENNET-CLARK.

Network neology

There has been much interest in the literature recently in such circuit-elements as the nullator and the norator. The former has been facetiously defined as a one-port which is simultaneously open and short-circuit; the latter is a one-port sustaining arbitrary voltage and current. Combined, they give a two-port, the nullor, which is equivalent to a perfect operational amplifier.

It may be thought by the uninitiated that such elements are the thoroughly un-British products of foreign and fevered imaginations. But can we afford to ignore any part of scientific progress? Who knows but that in a few years the inventing of circuit elements may pass from just a sub-branch of technology to a fully competitive activity, perhaps to be included in some future Olympics. It is in preparation for such an eventuality that the following ideas are offered.

Four new circuit elements are proposed which as far as the writer is aware, have not been previously described. They are the notator, the antiator, the unator, and the disator. Unfortunately no experimental versions have yet been produced. In the case of the first two this is understandable since they are both infinite-ports—a far more advanced concept than a mere one- or two-port element. A notator has an infinite number of input ports and if all carry different currents, the current in the output is *not* equal to any of them. Similarly an antiator has an output current which is the *opposite* of all the inputs simultaneously.

No definitions are provided for the unator and disator and these are offered freely in the hope that British genius can provide a solution. PETER WILLIAMS,

Paisley College of Technology, Renfrewshire.

John Clarricoats

On behalf of the 160 Meter Radio Amateurs, I wish to convey our sincere regrets and sympathies on the passing of John Clarricoats, G6CL. We shall all miss him greatly, for his enthusiasm and support of 160 Meter Band operations DXwise and propagation research, not only through R.S.G.B. but also recently through *Wireless World*. He was a fine gentleman in every sense of the word, as well as a dedicated radio amateur. STEWART S. PERRY, WIBB, Winthrop,

Mass., USA.



Gallium Arsenide Light Sources

Five gallium arsenide diode light sources from Texas Instruments Ltd., types OSX 1203, and OSX 1205-1208, emit near-infrared light when forward biased. Radiant output powers range from 20mW for the OSX 1205 to 200mW for the OSX 1208 at 25°C.

Types OSX 1203 and 1205 to 1207 are mounted in hermetically sealed packs with a flat window in the top of the case. The cathode makes electrical contact with the case and adjoining solder lug. The anode is in electrical contact with the stud which is insulated from the case by a glass-to-metal seal. Static forward voltage (V_F) is 2V maximum, spectral bandwidth with $I_F = 2A$, is 450 angstroms and emission band angle is 130°C. Diameters of the emitting crystals are 36 mm for the OSX 1203 and OSX 1205 and 72 mm for the OSX 1206 and 1207. The OSX 1208, which is mounted on a copper stud header for efficient heat dissipation has an output of 200mW at 25°C. The anode is in electrical contact with the stud, the cathode lead being a varnished 0.25mm copper wire fastened to the stud by a ceramic insulator. Spectral bandwidth and maximum static forward voltage are identical with the types OSX 1203 and 1207, being 450 angstroms, at half power points, and 2V respectively. Storage temperature of all devices is -55° to $+100^{\circ}$ C. These devices may be used in communications lines, infrared telescopes and binoculars, and in i.r. intruder detection systems. Texas Instruments Ltd., Manton Lane, Bedford, Bedfordshire.

WW306 for further details

Communications Receiver Assembly

Specifically designed for applications where continuous monitoring of fixed frequencies between 1.6 and 24MHz is undertaken, or for radio networks which have overlapping or simultaneous traffic schedules on a number of frequencies, Racal communications receiver, type RA1205, is said to provide a more economical solution to systems planning than continuouslytuned receivers. The complete assembly comprises eight receivers and an associated power unit. A double superheterodyne circuit is employed with crystals for the first and second v.f.o. housed in a temperature-controlled oven. The frequency range is covered in four switched bands, the operating frequency being selected inside the receiver during setting-up. As well as c.w. reception, a selection of filters gives suitable bandwidths for d.s.b., s.s.b. and f.s.k. modes of operation. The makers add that sensitivity, selectivity, intermodulation, image rejection and dynamic range all compare with the



world's foremost receivers. Racal Communications Ltd., Western Road, Bracknell, Berkshire. WW310 for further details

Versatile Power Supplies

Regulated variable power supplies designed to operate either as a constant voltage source or in a constant current mode are announced by Coutant Electronics. Voltage or current mode of operation is selected by a front-panel switch. The voltage is accurately set by coarse and fine controls and the current is adjusted by two similar but separate controls. Voltage and current levels are indicated on a built-in dual-scale meter. Thus it is possible after setting the supply to two different functions, to switch from constant-current to constantvoltage operation. In the constant-voltage mode



the output voltage remains within $\pm 0.005\%$ or 1mV of the preset level, whichever is the greater, for $\pm 10\%$ mains variation. Output impedance is less than 5m Ω at 100kHz; ripple voltage less than 1mV peak-to-peak and the transient response is such that on switching from no load to full load, the

output recovers to within 10mV of the steadystate voltage in less than 10µs. In the constantcurrent mode the input current is held to within $\pm 0.01\%$ or 0.1mA of the preset level, whichever is the greater, for a $\pm 10\%$ mains variation. The output current level varies less than 3mA from its setting for a change from zero to maximum output voltage. Operating temperature range is -10° to +45°C and the required input is 105-240V a.c. 45-400Hz. Three models are currently available: types LP.50/50 (0-50V d.c. 0.5A maximum); LP.100/30 (0-30V d.c. 1A maximum); LP.200/15 (0-15V d.c., 2A maximum). Each unit is housed in a case measuring 133 × 86 × 282mm deep and weighs 3.175kg. Coutant Electronics Ltd., 3 Trafford Road, Reading, Berkshire, RG1-8JR. WW 338 for further details

Hi-fi System

British Radio Corporation are about to enter the hi-fi market with a £200-300 ensemble comprising an f.m. stereo tuner/amplifier, record player and twin cabinet speakers. The usual controls are provided and tuning over the f.m. range 87.5-108.3 MHz is by five station selectors and press-buttons. Solid-state circuitry is used throughout employing 33 transistors and 15 diodes. Varicap diode type tuning is employed. On the audio amplifier side, socket facilities are provided for left- and right-hand speakers of $4-15 \Omega$ impedance, magnetic and ceramic stereo pickup input, tape recorder and auxiliary stereo input. Power output into a $4-\Omega$ load with sinewave input is approximately 15W with better than 1% distortion: into a $15-\Omega$ load, power output is approximately 10W. Frequency response measured at a constant output of 4W with bass and treble controls set level is 40Hz-16kHz ±1dB (30Hz-20kHz ± 3dB). Bass control range at 50Hz with reference to 1kHz is +12dB to -16dB as is the treble control range at 10kHz with reference to 1kHz. Approximate dimensions of the unit are 550mm long, 26mm deep, 100mm high. The record player employs a Goldring-Lenco GL75 transcription unit which has a 4kg turntable driven by a 4-pole constant velocity motor and has low wow and flutter levels. The pickup arm is provided with a hydraulic cueing device and an adjustable bias compensator. Speed adjustment is continuously variable between 15 and 18 r.p.m. and between 30 and 86 r.p.m. with click-stop positions for 16²/₃, 33¹/₃, 45 and 78 r.p.m. A lightweight interchangeable headshell is fitted with a Goldring 800/E magnetic cartridge. The twin speaker system employs Goodmans 305mm long-throw bass units and 102mm sealed back mid- and treblefrequency units fitted in cabinets with front face dimensions of 560×340 mm and a depth of 255mm. Crossover is at 1500Hz. Interconnections are made with connectors wired in the DIN configuration. British Radio Corporation Ltd., Thorn House, Upper St. Martin's Lane, London W.C.2. WW 339 for further details

Pulse Current Thyristors

Thyristors specifically designed for pulsed current operation are announced by Motorola. Six types, JAN2N 4199 to 4204 are especially applicable to military equipment such as pulse modulators for magnetrons, radar altimeters, surveillance and ranging radar, satellite systems and phased arrays. They are equally suited to similar civil applications. Forward current rating is 100A with peak forward blocking voltages between 300 and 800V. Maximum turn-on time is 400ns, di/dt rating 5,000A/µs and pulse repetition as high as 20,000 per second. Switching characteristics are stable over the temperature range -65° to 105°C and limits on all critical parameters are guaranteed. Motorola Semiconductors Ltd., York House, Wembley, Middlesex.

WW 337 for further details

I-band Solid-state Source

A compact J-band solid-state power generator type SSJ9 has been introduced by the M-O Valve Co. It is an electronically-tuned device suitable for local oscillator or test source use. The centre frequency can be pre-selected within the band 12.4 – 14 GHz with a tuning range of 250 MHz. Maximum power output is 5mW. The source operates from a 28V 100 mA supply and has an overall size of 85mm long × 37mm diameter. The M-O Valve Co. Ltd., Brook Green Works, London W.6. WW 322 for further details

Capacitor Bridge

A bridge designed for the measurement of all parameters of electrolytic and tantalum capacitors is announced by B.P.L. It is a four-terminal solidstate instrument, model CB154/4, which can be used with varying frequencies up to 20kHz and covers the capacitance range 0.01μ F to 1F. Leakage current measurement is provided in nine ranges enabling values as low as 0.1μ A to be detected. Another feature is a facility for the measurement of the voltage across the capacitor under test. Internal switchable bridge frequencies



and three separate panel meters are said to make for accurate reading on all parameters. Two internal bridge frequencies of 50Hz and 100Hz can be selected and an external frequency source up to 10kHz can be fed in. The internal polarizing voltage is fully stabilized over the entire range which is continuously variable up to 600V. Operation is from 115/125 and 200/240V, 50 or 60 Hz, a.c. The CB 154/4 measures $483 \times 305 \times 457$ mm and costs £425. British Physical Laboratories, Radlett, Hertfordshire.

WW 309 for further details

Portable Multi-band Receiver

One of the higher priced portable radio receivers now available is the Zenith Trans-Oceanic Royal 7000 which has been announced recently. This receiver carries a recommended retail price of £180 and provides reception on eleven wavebands. Refinements include a b.f.o., calibrated logging scale, and a log chart compartment and time zone indicator. The waveband ranges are: 150-400 kHz, 540 - 600 kHz, 1.6 - 9.0 MHz, 9.4 - 10.1 MHz, 11.4 - 12.3 MHz, 14.6 - 15.8 MHz, 17.1 - 18.5 MHz and 20.6 - 22.4 MHz. Bandspread tuning is provided on the s.w. bands. Also covered is the 88 - 108 MHz v.h.f. f.m. broadcast band and in the final position of the selector switch, the receiver is crystal-locked to the American v.h.f./f.m. weather broadcasts on 162.55 MHz. A "Norm-Sharp" i.f. switch modifies the i.f. bandwidth to reduce adjacent interference on s.s.b. and c.w. reception. Semiconductor complement totals 18 transistors (including the voltage regulator) and 9 diodes. A telescopic aerial is built-in for s.w. reception and a



ferrite rod provides for the reception of mediumand long-wave signals. The ferrite rod can be detached from the receiver and extended to a window to improve reception inside buildings. The receiver employs a 150mm elliptical loudspeaker and is powered internally by nine 1.5V cells, or alternatively it can be plugged into the mains supply via a fitted mains unit. The Zenith 7000 measures $240 \times 350 \times 160$ mm and weighs 7kg. U.K. distributors: United Mercantile Co. Ltd., Sovereign House, 13-14 Queen Street, Mayfair, London, W1X 8BB.

WW 320 for further details

Press-button Reed Switch

Contact bounce, a feature of press-button key switches which can be troublesome in keyboards associated with business machines and computer peripheral equipment, has been reduced by a new method employed by Starpoint Electrics. In their push-button reed switch, type 1RB1, two circular magnets are fitted; one to the switch housing and the other around the moving plunger. This gives a switching time of less than 1ms on operation. The 1RB1 is specially designed for p.c. boards and has terminals of pure nickel. It is claimed to have a life in excess of 10×10^6 operations. Keys may be mounted individually or in multiple arrays and complete keyboards with solid-state encoding can be supplied. Starpoint Electrics Ltd., 86 Coombe Road, New Malden, Surrey.

WW 321 for further details

Gunn Diode Oscillator

A stable and spectrally pure source of microwaves is provided by MI-Sanders' new Gunn diode oscillator, type 6061. It covers the frequency range 8.0 to 10.5GHz, and comprises essentially a hybrid mode cavity whose resonant frequency may be varied through the X-band by insertion of a micro-



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meter spindle, and a Gunn diode which is suitably mounted and shunted across the cavity. When biased between -8 and -16 volts, the diode will oscillate in a cavity-controlled Gunn mode at a frequency that is determined primarily by the micrometer insertion, and to a lesser extent by the characteristic properties of the diode. The tuning rate is approximately linear and an individual calibration chart is supplied with each oscillator. Microwave power (5mW typical) is coupled via an iris in the end wall of the cavity, which is matched at 9GHz by a fixed stub. Power output may be further improved by additional matching stubs in the outer waveguide. The instrument is priced at f70, f.o.b. U.K. A suitable power supply is the type 6590. Marconi Instruments Ltd., Sanders Division, Gunnels Wood Road, Stevenage, Hertfordshire. WW 333 for further details

Waveform Source

A waveform signal source which may be used in applications such as calibrating or evaluating oscilloscopes, measuring frequency response characteristics of voltmeters, measuring characteristics of low-pass filters etc. is announced by Britec Ltd. This is the Preston X-Mod 135 which provides three switch-selected waveforms: sine, triangle and square. Frequency is variable from 0.001 Hz to 10kHz in seven decades and output impedance is 50%. Output voltages are calibrated and range from 2mV to 10V peak via a 12-way switch and a 20:1 vernier control. Sinewave distortion is less than 1%, triangle linearity less than



0.5% and squarewave rise and fall time less than 1μ s. Construction is all-silicon with plug-in printedcircuit boards. Price: £209 plus import duty. Britec Ltd., 17 Charing Cross Road, London, W.C.2.

WW 313 for further details

Solderless Breadboards

Latest extension to the range of solderless breadboard systems by S.D.C. Products, are boards designed to accommodate integrated circuits as well as discrete components and to have an increased capacity of 208 contacts per board. One type (called µDeC) is intended primarily for i.cs and can accommodate two 16-lead d.i.l. or four 10-lead TO5 packages. The other type (called T-DeC) is for discrete components and can also accommodate one d.i.l. or two TO5 packages. The layout consists of two panels of parallel rows of electrically linked contacts spaced at 5mm which enables short lead devices to be inserted directly into the boards. The new boards may be interlocked to give an area of any desired size and each has slots to accommodate two control panels. Both types of board are supplied



in two packs. A single pack at f_2 10s ($f_2.50$) for T-DeC and f_2 15s ($f_2.75$) for μ -DeC contains one board, one control panel (with bushes for reducing the diameter of drilled holes in the panel) and a jig (for pre-forming components). A six-board pack contains six boards, six control panels, sets of bushes and jigs, fifty 1mm plugs and eight links for joining power rails in neighbouring boards. These cost f_15 for the T-DeC and f_16 10s ($f_16.50$) for the μ -DeC. S.D.C. Products (Electronics) Ltd., The Corn Exchange, Chelmsford, Essex.

General Purpose Op. Amp.

Type F1-8 operational amplifier, just added to the Comtec range, features an f.e.t. input. It is epoxy encapsulated in a shell measuring $31.8 \times 31.8 \times$ 15.2mm and has standard 7-pin configuration on 5.1mm centres. Operating from $\pm 15V$ supplies, the F1-8 has an input current of 50pA, an open loop gain of 80,000 and an output swing of $\pm 10mA$ at up to 50kHz. Protection is provided against supply reversal and short-circuit of any terminals. Computing Techniques Ltd., Westminster Bank Chambers, Bridge Street, Leatherhead, Surrey. WW 335 for further details.

Edge Connector

Designed for use with standard 1.6mm thick printed circuit boards having a contact pitch of 0.38mm a new edge connector is announced by Mullard. The bifurcated contacts ensure low circuit resistance and good reliability even when used in conjunction with misaligned or warped boards. Reliability if further enhanced by a heavy goldplated finish over the whole of the contact spring. The connector, type 036, is available with up to 45 contacts for single-sided or 90 contacts for doublesided boards. The body moulding, of black synthetic resin, is designed so that a connector can easily be cut into shorter lengths for experimental purposes. The loose fixing feet simply clip on the the ends of the connector body. Mullard Ltd., Torrington Place, London W.C.1.

WW 336 for further details

Digital Voltmeter

A compact digital voltmeter employing t.t.l. integrated circuits, an f.e.t. input amplifier and quartz crystal clock is announced by Verus Electronics. It is the company's type DC4500 and it features data output, isolated input, remote control and command



ranging. Storage circuitry incorporated into the drive to the neon display tubes ensures a nonflickering display. Decimal point indicators are positioned by the range switch. Readings may be initiated externally, manually or internally at rates up to 20 readings per second. Polarity and magnitude of the input are determined in one measurement, using a single zener reference source. Overload is indicated by a full-scale reading of 4999 with correct polarity indication, the accuracy of subsequent readings being unimpaired. Measure-ments from 0.1mV to 499.9V d.c. in four ranges are possible with an accuracy of 0.1% and resolution of 0.02%. Input impedance is 10Ω on all ranges. The instrument measures 228 × 120 × 254mm and weighs 4kg. Price £347. Verus Electronics Ltd., 122-124 Charing Cross Road, London W.C.2. WW 329 for further details

Small Sound-system

A radio/intercom system kit by Emerson-Rittenhouse comprises a master station (illustrated) with a.m./f.m. radio, three indoor remote stations, and an outdoor remote station, plus an installation kit of 36m of 5-core cable, plaster frame, power transformer, and a.m. and f.m. aerials. The equipment provides two-way communication from any position and it will relay radio programmes, tapes or records. It also enables the door or telephone to be answered remotely. The door-bell push overrides



any service in operation with an optional electronic chime. The system is transistorized and can be extended up to a maximum of eight remote stations. U.K. agent: Van den Bosch Ltd., Europair House, Alexandra Road, London S.W.19. WW 325 for further details

F.E.T. Pairs with S-clip

Matched pairs of field effect transistors type BFS21 and BFS21A from Mullard are now supplied with a new-style "temperature equalizer". This is an S-shaped clip with thermal characteristics similar to those of the rectangular heatsink previously supplied. However, as the transistors in the clip can be easily removed or re-orientated, the new S-clip makes possible greater flexibility in circuit layout. The matched f.e.ts in each pair have an extremely high input impedance (10%), low feedback capacitance (0.75pF), and a very low noise voltage (7.5nV at 10kHz with bandwidth of 5Hz). When in the clip, the transistors are maintained at the same temperature and the differences between their gate-to-source voltages do not change with changing temperature by more than $75 \mu V/deg C$ (type BFS21) or $150 \mu V/deg C$ (type BFS21A). Mullard Ltd., Torrington Place, London W.C.1. WW 326 for further details

I.C. Accessories

Two Augat i.c. accessories being marketed in the U.K. by Electrosil are an i.c. breadboard and test panel, and a socket removal tool. The 8130 series breadboard is for dual-in-line i.cs and takes up to 50 packages. Solderless interconnection is used





throughout and sockets have large contoured entry holes for easy i.c. insertion. Wiping goldplated contacts are employed. The removal tool enables socket bodies to be withdrawn from the breadboard panel in order to replace damaged or broken wire wrap contacts. The tool lifts the socket body from the board, and when the damaged contact is replaced, the body can be pressed back into place. Electrosil Ltd., Pallion, Sunderland, Co. Durham.

WW 331 for further details

High Value Capacitors

Sprague announce an extension to their range of 36D "Powerlytic" capacitors in a 76mm diameter \times 220mm long can which, they say, gives the highest C/V product available in a single electrolytic capacitor. Values available in this can size range from 650,000 μ F at 3V d.c. working with a maximum e.s.r. of 0.0120 Ω and maximum ripple current at 120Hz and 65°C of 26.7A, to 15,000 μ F at 150V d.c. working with a maximum e.s.r. of 0.0240 Ω at a maximum ripple current at 120Hz at 65°C of 18.9A. Sprague Electric (U.K.) Ltd., Trident House, Station Road, Hayes, Middlesex.

I.C. Voltage Regulators

Motorola i.c. voltage regulators in a TO-66 package delivering up to 500mA without the use of external power transistors are announced by Celdis Ltd. Using a single external power transistor, the load current can be boosted to more than 10A. Electronic "shut-down" and output short-circuit protection features are built-in. Input regulation is of the order of $0.002\%/V_{IN}$ and published data sheets specify output impedance. The TO-66 encapsulation has a 10-W power dissipation up to 65° C. Three types available are MC1460G, MC1460R and MC1560G, priced at f_{2} 8s 3d $(f_{2}.41)$, f_{3} 2s $(f_{3}.10)$ and f_{10} 6s 6d $(f_{10.32\frac{1}{2}})$ respectively. U.K. agent: Celdis Ltd., 43/45 Milford Road, Reading, Berkshire.

WW 314 for further details

Digital Picoammeter

A digital picoammeter by Keithley Instruments, model 445, measures currents over nine ranges from $10^{-9}A$ full scale to $10^{-2}A$ with a resolution of $10^{-12}A$. It features automatic polarity and overload indication and overload protection up to 1000V. Range selection can be manual or automatic. The circuit uses an f.e.t. input followed by a differential transistor amplifier stage and a transistor output stage. The analogueto-digital convertor is a dual-slope integrating



type composed mainly of integrated circuits. Line frequency rejection is 60dB with filter out; 100dB on the 10^{-9} to $10^{-7}A$ ranges with filter in. Common-mode rejection is such that a 100V d.c. or peak 60Hz a.c. signal will not affect the reading. Time stability is better than 0.5% of full scale per week and temperature stability is better than 0.05% of full scale per deg C. Offset current is less than 10-13A. Display rate may be adjusted from 24 readings per second to one reading per 10 seconds. Isolation of circuit earth from chassis earth is greater than 10° G shunted by 0.02 iF. Circuit earth may be floated up to ±100V with respect to chassis earth. Accuracy is from ± 0.5 to $\pm 0.2\%$ over the full measurement range. Power requirements are 105-125V or 210-250V a.c. in 50Hz or 60Hz versions, and dimensions are 130 × 482 × 254 mm. U.S. price \$1495. U.K. office: Keithly Instruments Ltd., P.O. Box 43, Reading, Berkshire.

WW 308 for further details

Programme-pin With Integral Diode

Although programming pins are available with provision for wiring-in the diode, Sealectro has now produced a component holder programme pin



which incorporates an integral diode, type 1S920. The diode is rated at 50V p.i.v., 200mA and is connected with cathode-to-tip polarity. A silverplated stem minimizes contact resistance. Sealectro Ltd., Farlington, Portsmouth, Hants. WW 311 for further details

Audio Equalizer

Full compensation in broadcast audio channels comprising treble and bass boost and cut together with mid-lift and cut at four spot frequencies is provided by EQ Series equalizers from Elcom. The transistor amplifier which has unity voltage gain is accessible for service on a hinged p.c. board. Input and output impedances are 600Ω unbalanced and output level is +15dBm maximum. Frequency characteristics are as follows: flat position + ldB 30Hz - 20kHz; presence ±10dB in 2dB steps at 1.4, 2.8, 4.0 and 5.6kHz; treble -15dB to + 12dB in 3dB steps; bass -15dB to +12dB in 3dB steps. Noise is -90dBm and distortion ranges from 0.03% at 0dBm output to 0.3% at +15dBm. Operating voltage required is 24V or 50V at 20mA. The equalizer front panel measures 177 × 62nm and it is 190mm deep. Elcom (Northampton) Ltd., Weedon Road Industrial Estate, Northampton. WW 328 for further details

Digital Tacho-ratiometers

Orbit 70 range of digital tacho-ratiometers by Orbit Controls of Cheltenham comprises a number of twin-channel frequency measuring instruments



with the added facility of measuring and displaying digitally the ratio between two speeds being measured as well as the absolute speeds. Instruments may be 4, 5 or 6 decade with timebase fixed at either 1 or 10 seconds, or with the timebase variable in lms steps from zero to 9.9995, allowing the display to be normalized to standard units (r.p.m., ft/sec., etc.). A counting rate of up to 106 seconds is possible. Input signals can be taken from any sensor which provides a pulse train of recurrent frequency bearing a direct relationship to the parameters being measured. Suitable sensors, including magnetic pick-offs, photo-electric detectors and photo-electronic tacho-generators are available from the makers. A b.c.d. output can be provided for control purposes or for the operation of printers or data-loggers. Orbit Controls Ltd., P.O. Box 16, The Runnings, Cheltenham, Glos. GL51 9PL.

WW 312 for further details

Thick-film D.C. Voltage Regulators

A comprehensive range of thick-film standard circuits has been introduced by Morganite Resistors Ltd. This range includes compatible positive and negative d.c. voltage regulators with output voltages from 3 to 32V. Built-in short-circuit protection and output voltage are adjustable by the addition of external resistors. The units have an operating temperature range from -55° to +125°C with a maximum load capability of 500mA, and a load regulation of ± 0.05%. Only 25.4 × 12.7 \times 4.3mm high, these units are manufactured using cermet thick-film passive elements fused to an alumina substrate, chip semiconductors are bonded to the substrates and attached to the passive elements by thermo-compression gold lead bonding. The addition of an alumina cover seals the unit. Regulators with a voltage range of 5 to 28 volts, load capability of 750mA and load regulation of 0.003%/mA are also available. Morganite Resistors Ltd., Bede Industrial Estate, Jarrow, Co.Durham. WW 330 for further details

Null Detector/Microvoltmeter

Sensitivity from $1\mu V$ to 1000V full scale is the main feature of null detector/microvoltmeter



announced by Keithley Instruments, of Ohio, U.S.A. The new transistor instrument, model 155, is battery operated. Input impedance varies from 1M C on the most sensitive range to 100M C and the makers add that there is "excellent immunity" from a.c. interference. An m.o.s. f.e.t. chopper is employed in the input circuit to provide good zero stability. Accuracy is $\pm 1\%$ of full scale at recorder output, 2% of full scale at the meter. Recorder output is ± 1V at up to 1mA. An overload feature allows up to 1200V to be momentarily applied to any range. Normal mode rejection is such that an applied 50-60Hz signal 80dB greater than full scale peak-to-peak will not affect the reading. A common mode voltage 120dB greater than full scale will not affect the reading. Operating power is derived from four internally mounted zinccarbon batteries and a power unit accessory is available which permits operation from a.c. mains supplies. Model 155 measures 133×210×170 mm and weighs 2.7kg. Price \$325. U.K. office: Keithley Instruments Ltd., P.O. Box 43, Greyfriars Road, Reading, Berkshire.

WW 304 for further details

Designer's P.C. Board

Rapid construction of circuits at the design and prototype stages of development is provided by the "Tri-board" breadboard comprising a copperclad laminate, etched to leave rectangular lands of copper. Four continuous strips running the length of the board can be used as power supply and earth busbars and large strips at each edge serve for termination of input and output leads. Supplies are in glass fibre and s.r.b.p. laminates in grades suitable for cold punching or cutting.

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Board size is 190×140 mm, 1.6mm thick with 28g copper. Roller tinned finish is standard. The makers claim that the Tri-board may be usedover and over again. Price: 15s (75p) in glass fibre, 9s 6d (47 $\frac{1}{2}$ p) in s.r.b.p. laminate. Tric Instruments Ltd., "Allington", Dartford Road, Farningham, Kent.

WW 301 for further details

Avalanche Diode Sources

Available over the frequency range 5 - 18GHz new avalanche diode sources by Interplanetric contain built-in solid-state current control circuit: which require a 90V power supply and exhibit ar effective source resistance of $50k \Omega$. Additiona power drain on the supply is typically 250mW. The benefits of current control are improved amplitude and frequency stability with variations in powe supply, protection against reverse polarity and over voltage plus a reduction in incidental a. m. and power output uniformity from unit to unit. As adjustment is provided to change the power outpu over a moderate range without changing the supply voltage, the resultant frequency change bein typically 5MHz/mA. Centre frequency setting hold to within 0.5% over the temperature rang -40° to +70°C and second harmonic output i the 20-GHz region is typically 40dB down. Inter planetric, 39-49 Cowleaze Road, Kingston-upon Thames, Surrey.

WW 332 for further details



Unitor Pattern 103 DEF-5325-1 reporting for heavy duty!

MCMURDO UNITOR MICRONECTOR IS THE FIRST CONNECTOR TO GET MINISTRY OF DEFENCE APPROVAL TO DEF' SPECIFICATION 5325-1



Send for data sheet and further information to: The McMurdo Instrument Co. Ltd. Rodney Road, Portsmouth, Hants. Telephone 35361 Telex 86112 Ministry Approval Certificate 1693 was the honour earned by McMurdo Unitor Pattern 103 – the toughened version of the universally trusted Micronector.

With specially treated contacts and D.A.P. moulding the Unitor is ready to do battle with extremes of environmental conditions.

It is polarised for simple, trouble-free connection and is available with a range of covers and masks similar to those supplied for the Micronector range. The full range of 7, 14, 26, 34, and 50-pole versions of the new Ministry of Defence approved Unitor connector are available.

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Authorised Stockists :- LUGTON & CO.LTD., 209/210 Tottenham Court Road, London W.1. Tel: Museum 3261. SASCO, P.O. Box No. 20, Gatwick Road, Crawley, Sussex. Tel: Crawley 28700 (also Chipping Sodbury 2641, Cumbernauld 25601 and Hitchin 2242) and agents in principal overseas countries

WW-112 FOR FURTHER DETAILS

Wireless World, May 1969





CDU 150 is a fully transistorised general purpose versatile oscilloscope with many advanced features and extremely high engineering quality.

- Bandwidth DC 35 MHz at 5 mV/cm -Dual Channel.
- Large (8 x 10cm), bright display.
- Measuring accuracy \pm 3% 'X' and 'Y' all ranges.
- Full delayed timebase facilities with gated mode.
- Compact, rugged, lightweight.



 $\pounds470$

The Cossor CDU 150 has been selected, in open competition, from a wide range of commercial oscilloscopes by the Joint Departmental Radio and Electronics Measurements Committee to be the general purpose oscilloscope for use by the British Armed Forces and associated Ministry Departments. The CDU 150 satisfied the exacting requirements of this committee in respect of electrical performance, construction, and ease of operation and maintenance.

Take a good look at the Cossor CDU 150 - send for comprehensive literature.

Cossor Electronics Limited, Cossor Instruments (Industrial Products Group) The Pinnacles, Elizabeth Way, Harlow, Essex, England Telephone: Harlow 26862 Cables: Cossor Harlow England Telex: 81228

WW-113 FOR FURTHER DETAILS

www.americanradiohistorv.com

Oscilloscope Trolley

An instrument trolley made in a variety of versions to suit almost every known make of oscilloscope is announced by Avoncel. It costs $\pounds 25$ and can be ordered by simply stating the make and model number of the oscilloscope for which it is required. Of welded construction, the framework (25.4 x 25.4 mm) is of 14 gauge mild steel with 14 and 16



gauge mild steel shelves. Good stability is claimed with 113kg load. Features include a fail-safe top shelf adjustment which enables the viewing angle to be adjusted from $0-20^{\circ}$. Both top and bottom shelves have lipped edges and the top shelf is fitted with adjustable back stops. Four-inch swivel castors are fitted; the two front castors are equipped with brakes. A four-way 13A power distribution box is fitted. Avon Communications and Electronics Ltd., 318 Bournemouth (Hurn) Airport, Christchurch, Hampshire.

WW316 for further details

Transient Voltage Suppressors

Two new types, designated the KSA and KSL series, have been added to the range of Klip-Sel transient voltage suppressors by International Rectifier. These units are designed to give protection to electrical equipment and systems which are subject to voltage transients and the new series utilizes a new type of selenium plate offering improved reliability and reduced steady state power loss. International Rectifier, Hurst Green, Oxted, Surrey.

WW319 for further details

11-in Television Monitor

Picture monitors for use in industrial and educational television systems have been introduced by E.C. & S. of Newark. They are models 11T and 11T/S and are especially suitable for mobile duties because of their compactness, light weight



and mechanical rigidity. Silicon transistors are employed throughout and an 11-in rimguard c.r.t. Construction is of four printed circuit panels with heavy components mounted on the unit chassis. Both composite and non-composite video signals will be accepted and these may be bridged or terminated by the operation of toggle switches on the rear panel. Model 11T/S may be fitted with an additional optional circuit module producing random mixed syncs, which are fed from the monitor via a coaxial connector on the rear panel. This facility allows the outputs from several cameras to be fed into a singled monitor via a video switching unit without incorporating an independent synchronizing pulse generator. The random mixed syncs with frame pulses locked to the mains may be used to lock the associated cameras, providing stable displays on camera switching. Prices: £195 (11T) and £205 (11T/S). Electronic Control and Surveillance Ltd., Queens Head Court, Newark-upon-Trent, Notts.

WW 302 for further details

25kV E.H.T. Unit

An e.h.t. unit suitable for providing the 25kV e.h.t. supply to most projection television receivers and for conventional shadowmask tubes when separate circuits are used for line deflection and e.h.t. is announced by Valradio. The circuit recommended for use in conjunction with the e.h.t. unit includes a feedback regulator to provide zero output impedance between no load and the maximum current of 350 μ A. The 25kV is smoothed



and fed via a 700mm screened cable to an outlet plug to fit Mullard c.r.t. type MW2. Power supply requirement is 350V at up to 70mA at an effective impedance of 400-600 Ω Size of the e.h.t. can is 107 \times 107mm with an overall height of 127mm including the top connections. Valradio Ltd., Browells Lane, Feltham, Middlesex. WW 305 for further details

Flexible Switch Innovation

Described by Plessey as a "new concept in rotary switch techniques" their new PrintSwitch features a unique printed pattern which enables any contact configuration to be obtained by removal of unwanted inter-connections. Shorting or non-shorting contacts on a single pole can easily be arranged. To this has been added better standards of reliability by the elimination of contact alignment problems associated with conventional clipped wafer switches. The PrintSwitch uses a printed epoxy glass stator assembly and rotor contact pressure is obtained from low-rate coil springs. Dual-compression springs and rollers are used for indexing. All con-





nections are provided in a single plane at standard pitch so that direct insertion into a printed board is possible. Metric spindles and bushes are standard and the usual facilities such as concentric shafts are also available. Professional Components Division, Plessey Components Group, Titchfield, Hampshire.

WW380 for further details

Tunnel Diodes

New miniature tunnel diodes announced by Mullard have a switching time of one nanosecond and are primarily intended for use in the switching circuits associated with ferrite-core memories. The diodes, type AEY25 and AEY26, are in DO-17 encapsulations that are only 2.54mm long and 2.8mm in diameter-small enough to be mounted around the edge of the matrix plane. Consequently, very short connecting leads can be used and high operating speeds are achieved. Operating characteristics are: maximum forward current 50mA; typical peak current 4.7mA (AEY25), 5mA (AEY26); peak voltage 75mV; and valley voltage 330mV. Ratio of peak current to valley current (I_P/I_V) is typically 6. Mullard Ltd., Torrington Place, London, W.C.1.

WW381 for further details

Broadband Miniature Amplifier

A new low-noise transistor amplifier by Electro/ Data Inc., U.S.A., covers the entire v.h.f. and and u.h.f. frequency bands and is contained in a shielded package no larger than a normal coaxial connector with 7-mm input and output plug-andsocket features. It has input and output impedances of 50Ω (nominal) and requires only a single negative 18V, 20mA d.c. power source. The amplifier, model A10, has a 10dB gain response from 20-1000MHz and greater than 18dB from 50-850MHz. Increased gain can be obtained by cascading two or more units. Noise figure is 4dB (typical), 5dB (max). Weight of the unit is 21g and the price for single quantity \$300. Electro/Data Inc., 3121 Benton Street, Garland, Texas 75040, U.S.A. WW315 for further details



Letter from America

Here is a success story that is worth recording. Back in 1939, two electronics engineers decided to go into business for themselves and so they began in a garage with a total capital of about \$500 (then worth about £120). At first they made a few diverse and exotic products but then moved into the instrument business with an audio generator. This was an RC device and eight of these instruments were used by the Disney studios for sound effects in the famous "Fantasia" film. Soon the garage became too small as more and more instruments were added to the range. Such was the phenomenal rate of expansion that by 1956 annual sales had sky-rocketed to \$20 million, reaching the incredible figure of \$200 million by 1966 and \$265 million last year. As you have probably guessed, the two engineers are William Hewlett and Dave Packard. The firm now employs over 13,000 workers all over the world and Dave Packard is now Deputy Secretary of Defense in the Nixon Administration. This appointment has met with some criticism on the basis of HP's large business deals with the U.S. Government. Although Dave Packard has resigned from the company and placed his \$300 million stock in a charitable trust the critics say could he really be impartial to his own firm even if he sold all his stock? On the other hand, his supporters argue that the government should not be denied the services of so talented and experienced an executive just because he is successful!

There are now nearly 20 million American households with colour TV-something like 33% of the total. Sales of colour receivers increased more than 10% in 1968 to 5,771,000 of which 800,000 were imported from Japan. The import situation is even worse in Canada where the influx of Japanese receivers increased from 3.7% in 1965 to more than 25% at present. The president of the Canadian Electronic Industries Association said recently that no less than 56% of all TV sets with 19-inch screens or smaller are imported from Japan. About 190,000 colour sets were sold in Canada last year and the 1969 forecasts are up to 250.000.

Curiously enough, varactor diodes are not used as tuning elements in American television sets. This is in sharp contrast to European practice: if I remember correctly Grundig used them as far back as 1965. The reason is mainly due to technical difficulties arising from the large number of channels involved. Manufacturers must provide a choice of 12 channels in the v.h.f. band plus another six in the u.h.f. range. So selectivity becomes a problem and there is the added cost of the switches to contend with.

Infra-red systems for night vision (Snooperscopes!) were used in World War II and I believe some are still employed in Vietnam. Now a much better method has been devised using lasers. The complete system comprises a laser diode, optical assembly, infra-red tube, modulator and delay circuitry and a 28-volt battery. Coherent light from the gallium arsenide laser is pulsed to the target and the tube is cut-off until the ray returns to prevent ambient light from entering. The reflected pulse switches the tube on and the viewer sees the target image on the lens. To see moving objects the instrument can be switched to a d.c. mode and then when once located the object can be watched with greater definition by reverting to the pulse mode. Range is said to be up to 300 feet and the weight of the whole outfit is 15 1b. It is priced at \$3,000 (£1,250) and the maker is Laser Diode Laboratories, Inc.

Westinghouse have developed a TV pick-up tube which if not capable of seeing in the dark, will certainly operate with extremely low illumination. It uses the principle of secondary emission (s.e.c.) and has a rated sensitivity of 20,000 microamps per lumen. They are used in the colour cameras for medical purposes, electron microscopes, and industrial and military applications where a good resolution under low lighting conditions is required.

General Telephone & Electronics Corporation are also working with lasers and they disclosed recently that they had evolved what they termed a "technological breakthrough" in the use of lasers for large-screen colour TV. Picture sizes up to 48 by 31 inches have been achieved although it was emphasized that the system is still very much an experimental one. The system uses two lasers to produce three laser beams. A krypton gas laser provides red light and an argon laser gives blue and green light. The beams pass through electro-optical modulating equipment where signals from a standard colour receiver are impressed on them. An arrangement of mirrors then combines the three beams into a multicolour beam which travels to a prism that splits it into a pair of similar full-colour beams. The two fullcolour beams are then directed by mirrors to a rotating 15-sided mirror which scans them in rapid succession, producing the horizontal lines needed for a TV picture. The scanned beams are then reflected to a vibrating mirror that produces vertical motion. From there the light rays are reflected to a large screen where the picture is produced.

Ultrasonics are used for all kinds of things these days and recently I came across ultrasonic bird repellers. These are made by a firm rejoicing in the name Bird-X of Chicago and work in the 20-kHz range. One model uses a transistor oscillator and the other is operated by compressed air with focusing reflectors.

The 59th annual Boat Show held recently in New York attracted a very large attendance and exhibitors were very optimistic about the future. With good reason-sales now top \$15,000 million and there are now more than 8 million boat owners in the U.S Electronic devices were well in evidence and one of the most interesting was a "wrist range finder" made by Hartmann. This gad get is a micro-miniature depth sounder worn like a wrist watch for skin divers. It has a range of 360 feet and has been tested to withstand submersion to 220 feet. The device can tell the wearer the distance to the bottom, to the surface, to underwater obstructions or to other swimmers. The price? \$129.99 (say £50). Dozens of radic telephones were to be seen but here the situation is complicated by an F.C.C. regula tion which states that January 1971 is the latest date for the installation of a.m. transmitters using the 2-3 MHz marine band.

However, sales of the new more expensive f.m. units have been relatively slow and the actual figures show less than 200 f.m. unit. out of a total number of some 167,000. One of the reasons might be due to some confusio. over F.C.C. cut-off dates and also whether amplitude modulated s.s.b. will be permitted This state of affairs is not entirely due to F.C.C. vacillations but rather to the necessity of co-ordinating with the requirements o the International Telecommunication Union The manager of Raytheon Marine Product. Division said that sales of v.h.f./f.m. radio telephones have yet to capture a mass marke because of their higher price and to an under standable inertia on the part of the genera public until definite time schedules ar announced by the F.C.C. He went or to say that he expected that the final dat for the use of a.m. radio telephones excep those employing s.s.b. would be 1977 afte which only v.h.f./f.m. equipment could b used.

Radar may well be standard equipmen on millionaires' yachts, but it is still luxury item for most boat owners, neverthe less, prices are coming down. For example Kelvin Hughes had a radar system at les than \$2,700 (£1,100) and small boat system were also shown by Decca and Astaron Birc Some of the larger, expensive boats were no only equipped with the latest navigation: aids but boasted wall-to-wall carpeting, TV stereo and a full complement of washin machines, dishwashers, and other gadgets i the galley-sorry-kitchen! About the onl thing missing were Bird-X devices to keep th G. W. TILLETT sea-gulls off!

Public Address Equipment

The A.P.A.E's Coming-of-age Show at Harrow

This small but exclusive annual exhibition of public address equipment, arranged under the auspices of the Association of Public Address Engineers, was again held at the King's Head Hotel, Harrow, from March 11th to 13th. This year's show marked the Association's 21st anniversary. The exhibition of equipment which ranged from large custom-built desk console systems to handy personal p.a. kits was supported by a series of lectures giving to the show an atmosphere of an annual swopping of ideas. Transistors and printed circuits have enabled the modular type of construction to be employed almost invariably in new equipment so that versatile installations can be achieved using a minimum of components by exchanging only the significant section of the total circuit. Because of the arduous conditions in which p.a. equipment is expected to operate, most of the housings are robustly constructed, and in some cases speakers were proofed against corrosion and even against flame. Public address work, mobile p.a. in particular, can be an untidy business requiring the plugging and unplugging of ancillary units between installations and "hooking-up" a variety of configurations of microphone and music inputs, speaker outputs and so on, to satisfy the requirements of individual functions. Most manufacturers have commendably adopted the DIN specification for their inputs and outputs but the p.a. equipment handler may be tempted to question the ability of the DIN-type plug and socket to withstand the inevitable rough handling it will receive.

One welcomed innovation at this year's show was the introduction by the A.P.A.E. of a 45 r.p.m. test disc costing 15s (75p). This is the first time that a facility of this sort has been available to the p.a. engineer and with it he can apply a more professional touch to the testing of his installation. Instead of the "one, two, three . . . testing", kind of exercise the installer will simply start the test record then check his equipment for volume levels, speaker phasing, and correct functioning generally. The first side of the record contains three minutes of male speech ending with 30 seconds of talking away from the mic, followed by three minutes of female speech. The second side contains selected test phrases followed by a 1-kHz tone, a warble tone, and pink and white noise.

Of the equipment being shown, the largest

display was a sophisticated desk console by Audix which, although appearing in a standard version, can be varied to accommodate different combinations of sound and control equipment. Consoles are normally custom designed and are intended for lecture theatres, concert halls and the like. Combinations include microphones, tape recorders, gram turntables and radio tuners with the usual level controls on each channel, output metering and circuit switching. The output is suitable for feeding via G.P.O. lines to remote power amplifier and speaker distribution networks. At the other end of the scale there was what might be described as a one-man public address system, the Bouyer Clubflex 201. This is a French import shown by D. A. Lyons and Associates. It comprises a collapsible microphone stand to which is clamped a quadruple unit column-type loudspeaker and reading lectern. The speaker unit contains a 5-W amplifier and eight U2 batteries. The equipment features a dynamic hand microphone which can be removed from the stand and has a reasonable length of connecting cable to allow the talker to move about if required. The Clubflex 201 weighs 3.9kg and costs 64gn (£67.20). Ultra Electronics Ltd. were



Philips stack-unit p.a. assembly

featuring their TAM25 and TAM50 amplifiers which are basically power output stages of 25-and 50-W rating respectively. The amplifiers are designed with a separate section on the chassis containing just six edge connectors for the insertion of plug-in p.c. boards. Individual boards are available for microphones, radio, records, tape, etc., thus enabling the role of the installation to be changed quickly. A further six inputs could be added to the system by means of a separate audio mixer. One permutation of modular assembly is simply to stack units one on top of the other as required, a method favoured by the Philips Division of Pye TVT. The cases are designed to fit one on to the other in building-brick fashion and thus, using three basic amplifiers of 25-, 50- and 100-W output and three types of preamplifier, nine different combinations are possible. For future extension additional units can be added to the stack including a radio tuner, record player, tape recorder or signal switching unit.

Microphones to be seen at the show were many and varied but there was a handful of new types. There was, for instance a Reslo dynamic microphone type SL1 in the highquality bracket at a cost of $\pounds 38$. It has a frequency response of 100-16000Hz and sensitivity 84dB below 1V/dyne/cm². Available in low- or medium impedance versions, response is omni-directional.

Another new range of microphones was the Series 4 by Lustraphone. Of these the 4-10 and 4-11 are respectively desk and lavalier type using the same microphone head. Frequency response is substantially flat from 200 Hz to 11kHz with an omnidirectional pick-up pattern. Output level is -86dB at 50Ω at 1kHz (0dB = $1V/dyne/cm^2$); high- and low-impedance versions are available. No dramatic developments were to be seen in loudspeaker design but one feature worthy of mention is the use of glass fibre for the construction of re-entrant horns in place of metal. This material is used in the construction of Rola Celestion's model FG and it makes the speaker impervious to salt air on boat decks or chemical atmosphere in some industrial plants. Rola Celestion is not the first speaker manufacturer to use glass fibre construction but it does no harm to be reminded that the electronics industry generally, probably more than any other industry, takes every opportunity to exploit the benefits offered by new materials.

An important aid to audio engineers, particularly in locations where high ambient noise is troublesome, is a new portable sound level meter (model SLM3) by Acos. The SLM3 complies with BS 4142, making it suitable for the measurement of industrial noise nuisance and BS 3425 for measurement of vehicle noise, so that its use could widen the scope of p.a. dealers' activities. Fast and slow meter damping facilities are provided (peak and average levels) and an output socket enables the meter to be connected to a pen recorder. The sound level meter comes in an attache case kit which includes accessories to extend the range and scope of the basic instrument, comprising an amplifier module, calibrator unit, windshield, comparator unit (for differential measurements), tripod and extension lead.

Test Your Knowledge

Series devised by L. Ibbotson* B.Sc., A.Inst.P., M.I.E.E., M.I.E.R.E.

12. Acoustics

1. At a point in free air carrying a plane travelling sound wave the excess pressure variation with time is observed to be sinusoidal. The particle velocity at this point will

- (a) be constant
- (b) be constant in magnitude, but change in direction when the excess pressure changes sign
- (c) vary sinusoidally with time in phase with the pressure

(d) vary sinusoidally with time in phase quadrature with the pressure.

2. The velocity of an audible sound in free air depends on

- (a) the total air pressure
- (b) the absolute temperature
- (c) the frequency
- (d) the intensity.

3. A sound source moves towards and away from the observer in a cyclic manner. The observed frequency

- (a) is unaffected
- (b) is increased
- (c) is decreased
- (d) rises and falls cyclically.

4. The overtones produced by a closed organ pipe are

- (a) inharmonic
- (b) all harmonics of the fundamental
- (c) odd harmonics of the fundamental
- (d) even harmonics of the fundamental.

5. On a piano (tuned in equal temperament) middle C is found to have a frequency of 260Hz, the C above it 520Hz. An octave has 12 semitones. The frequency of the note 6 semitones above middle C (F#) will have a frequency

- (a) $260 \times \frac{3}{4}$ Hz
- (b) $260 \times 6\sqrt{12}$ Hz (c) $520 \times \sqrt{12/6}$ Hz (d) $260 \times \sqrt{2}$ Hz.

6. Two pure sound tones, one at 100Hz the

other at 1000Hz, sound equally loud. The threshold of audibility is 10^{-12} watt per square metre at 1000Hz, 10^{-9} watt per square metre at 100Hz. The loudness levels of the two signals in phons

- (a) are the same
- (b) differ by 3dB
- (c) are in the ratio 3:1
- (d) are in the ratio 1000:1.

7. The dynamic range of human hearing at frequencies around 1kHz is of the order of (a) 10dB

- (b) 60dB
- (c) 130dB
- (d) 200dB.

8. The mechanical response of the moving parts of a loudspeaker or microphone can conveniently be analysed using methods which are analogous to those of electric circuit analysis. If force is regarded as analogous to voltage the mechanical analogue of inductance is

- (a) inertial mass
- compliance (the reciprocal of stiffness) (b)
- (c) viscous friction
- (d) velocity.

9. The moving parts of a loudspeaker have at a given frequency a mechanical impedance Z_m . The extra electrical impedance introduced by the motion of these parts (the motional impedance) is

- (a) directly proportional to Z_m
- (b) directly proportional to $Z_m^{m_2}$
- (c) inversely proportional to Z_m^2 (d) inversely proportional to Z_m^2

10. The fundamental resonant frequency of a cone loudspeaker should be

(a) below the lowest frequency which it is required to radiate

(b) in the centre of the range of frequencies which it is designed to radiate

(c) above the highest frequency which it is designed to radiate

(d) outside the audio range of frequencies.

11. A simple cone loudspeaker becomes increasingly ineffective as the frequency of the sound which it is required to radiate increases above a value corresponding to a wavelength equal to the circumference of the cone. Select the factor below which is not relevant

(a) the radiation from the cone becomes increasingly directive

(b) there is a fall off of total radiated power

(c) the cone "breaks up"

(d) cancellation can occur between radiation from the two sides of the cone.

12. An exponential-horn loaded loudspeaker can give uniformly efficient energy conversion over the whole of the audio frequency range provided the mouth is sufficiently large. The horn must, however, be long because the taper must not be too sharp. The disadvantage of a short horn with a wide mouth is

(a) it exhibits resonances at low frequencies (b) it exhibits resonances at high frequencies

(c) it has a high cut-off frequency below which no energy is radiated

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(d) it has a low cut-off frequency above which no energy is radiated.

13. Of the following types of microphone one requires an applied d.c. potential

- (a) a crystal microphone
- (b) a capacitor microphone
- (c) a moving-coil microphone
- (d) a ribbon microphone.

14. A ribbon microphone and a crystal microphone (both normal of their type) feed amplifiers which are adjusted so that the output is the same from each when the two microphones are exposed to a plane sound wave in free air. Both microphones are placed in a large tube, which supports a total standing wave of sound, at a pressure antinode. The output will be

- (a) very small from both
- (b) very large from both
- (c) large from the ribbon microphone, small from the crystal microphone

(d) large from the crystal microphone, small from the ribbon microphone.

15. The basic polar diagram of a normal ribbon microphone is

(a) figure-of-eight at all frequencies

(b) omnidirectional at low frequencies;

figure-of-eight at high frequencies

(c) omnidirectional at high frequencies;

- figure-of-eight at low frequencies
- (d) omnidirectional at all frequencies.

16. It is normally accepted that the optimum reverberation time for a small room in which music is to be reproduced should

(a) be the same for all frequencies

(b) decrease uniformly with increasing frequency

(c) decrease with increasing frequency up to about 1kHz and remain constant for all frequencies above this.

(d) be constant with increasing frequency up to about 1kHz and decrease with increasing frequency above this.

Answers and comments, page 247

"Test Your Knowledge—11"

I must apologize to readers for question 6 of No.11. The answer which I gave is incorrect, and the alternative solutions given in the question are inappropriate.

R.F. amplifiers of the sort abown are generally designed on an impedance matching basis for maximum power transfer. Thus the input impedance of the next stage will be transformed to such a value that it is presented to the transistor as equal to its output impedance (Losses in the coil can generally be neglected). With the usual values of input and output impedances a parallel tuned circuit connected directly across the transistor would require, to give the optimum bandwidth for amplitude-modulated sound reception, values of inductance and capacitance which are inconvenient; the capacitance value required is very large. A smaller value of C can be used if a larger tapped coil is used in the way shown so that the required values of L and C are presented to the transistor. The transformed impedance of the next stage presented to the transistor must still be the same.

If, with a stage designed in this way, the collector connection were moved to point Q on the diagram, the bandwidth would be too wide, not too narrow as suggested by my answer.

I am indebted to Mr. M. G. Scroggie for drawing my attention to this fact.

^{*}West Ham College of Technology, London, E.15.

Wireless World, May 1969

World of Amateur Radio

Famous Contact Recalled

A link with perhaps the most famous amateur contact of all time-the first twoway, short-wave, transatlantic contact of November 28, 1923-has been broken with the death on January 21 of Leon Deloy ex-8AB of Nice, France. It was Deloy, more than any other European, who was responsible for starting the rush to shorter wavelengths. In the Spring of 1923 he began experiments on about 100 metres and soon formed the opinion that these wavelengths could "render immense and unsuspected service in long distance work". During that summer he visited the United States and arranged with Warner, Schnell and Handy of A.R.R.L. to try such wavelengths in the 1923 transatlantic tests. On his return to France Deloy began a long series of tests with E. J. Simmonds, 20D of Gerrards Cross. In late November, he cabled A.R.R.L. asking them to listen for him on 100-110 metres. Fred Schnell, 1MO, received strong signals from him on November 27, and the following night two-way contact was made with Schnell reporting 8AB's signals as "U ALSO VY QSA TWENTY FEET" indicating that he estimated the signals were strong enough to be heard 20ft from the headphones! Almost immediately contact was also established with John Reinartz, 1XAM, and the historic three-way contact continued for several hours. As news of this success became known, the great rush by other amateurs and many commercial communications organizations to exploit the shortwaves gained momentum.

New Award Stimulates H.F. Activity

The level of DX activity on the main h.f. bands seems to have been stimulated by a new A.R.R.L. "Five Band DX Century Club Award". This sets amateurs an extremely tough requirement of contacting 100 different countries on each of five h.f. bands: 3.5, 7, 14, 21 and 28 MHz, and only contacts made after January 1, 1969 are counted. Considerable Commonwealth activity was also noted during the annual B.E.R.U. Contest, organized by the R.S.G.B., during early March and held this year under good propagation conditions. Some of the overseas participants were heard giving contact serial numbers well over 500, and the 14 MHz long and short paths to Australia were open for many hours.

In general, however, many British

amateurs have become concerned at the lower level of h.f. activity apparent in this country in recent years, with a high percentage of licensed amateurs now apparently inactive on these bands. This is believed to be due in part to the continued problem of avoiding causing interference to television reception. It is hoped that more activity will be possible by using such techniques as absorptive low-pass filters (with which v.h.f. harmonic energy is separated from the h.f. power by means of cross-over filters and then dissipated in a resistive load) and various forms of ferrite transformers and baluns to reduce entry of fundamental and harmonic power into television receivers along the outer screen of coaxial feeder cables. Firm hope for the future is based on British television "duplication" plans since u.h.f. TV appears appreciably less susceptible to interference than Bands I and III, with Channel 1 notoriously suffering.

Long-distance V.H.F.

A new 144 MHz record is being claimed for a "moonbounce" contact between the Swedish station SM7BAE and New Zealand ZL1AZR. SM7BAE has also worked California on c.w., and has received s.s.b. from K6MYC. During March, the Rhodesian station, ZE1AZC on 50 MHz has been reported for the first time in the U.K., having been heard by G3JVL probably due to transequatorial propagation. Several good aurora DX openings on 144 MHz occurred during March. At an R.S.G.B. meeting, C. Newton, G2FKZ, put forward the view that, during aurora openings, longer east-west paths are possible at the beginning and end of such periods, and advised amateurs to continue operation in these conditions to about 02.00. He believes that "afternoon" auroral openings are produced by a different mechanism and do not give rise to such long distances. D. Hayter, G3JHM, predicted that "double hop" paths would be found feasible during sporadic E conditions, and that this could result in contacts between the U.K. and Middle East on 70 MHz. He announced that the Gibraltar beacon station ZB2VHF will soon be transmitting simultaneously on the 50, 70 and 144 MHz bands. Rhodesian 50 MHz beacons have been reported on a number of occasions in Gibraltar, where the South African beacon ZS6VHF has also been received.

Several v.h.f. and u.h.f. contests organized

by R.S.G.B. will be held during May, including 144 MHz portable (May 3-4); 432 MHz open (May 24-25); and 1296 MHz (May (24-25). G3GZL was winner of Section A of the 144 MHz c.w. contest held during January. G3VPK of Chelmsford led the field in the 70 MHz contest held during February. American amateurs recently regained the 2300 MHz record with a 225-mile contact using pulse techniques. Both stations were at about 1500 ft and it is believed the contact was effected by scatter propagation.

Gift to Mauritius Amateurs

The Johnson Viking Ranger II transmitter donated to the Mauritius Amateur Radio Society by the American Radio Relay League was recently handed over to the chairman of the society Paul Caboche (VQ8AD) during an informal ceremony at the United States Embassy. In making the presentation, William B. Hussey, the Charge d'Affairs, displayed a real interest in the society's training programme and disclosed his own earlier interest in amateur radio. News has been received that the construction of the transmitter is finished and the first QSO was made on 20 metres c.w. with station K6QPH/4 in South Carolina. It is hoped to install the transmitter in the society's clubroom and Region 1 of the I.A.R.U. has agreed to give assistance in providing a companion receiver.

Reciprocal Licensing

The current issue of the I.A.R.U. Calendar lists the names of 41 countries and their amateur societies, together with the names of those countries whose amateurs are accorded eligibility for amateur operating privileges when visiting that country. Information is given regarding the address from which forms and assistance in making application may be obtained. In the case of the United Kingdom, reciprocal licensing agreements have been signed with Austria, Belgium, France, Luxembourg, Netherlands, Morocco, Israel, Finland, Denmark, F.R. of Germany, Portugal, South Africa, Sweden, Switzerland, United States and all Commonwealth countries. Information can be obtained from the Radio and Broadcasting Department of the G.P.O.

In Brief: A special station GB2HRH will operate at Caernarvon from June 28 to July 6 to mark the investiture of the Prince of Wales. . . . Indonesia has notified I.T.U. of the withdrawal of objections to YB amateurs working other countries. . . . An I.A.R.U. Region 1 Conference is being held in Brussels between May 5 and 10. . . . The c.w. section of the U.S.S.R. "Peace to the World" h.f. contest takes place from 09.00 May 3 to 21.00 May 4, on all bands from 3.5 to 28 MHz. . . R.S.G.B. 1.8 MHz direction finding qualifying events are to be held at Stratford-on-Avon on April 27, Grimsby on May 18, Oxford on June 29, Salisbury on July 20, and High Wycombe on August 3. The national final is to be held at Rugby on September 21.... An American amateur was recently given a six-months prison sentence for transmitting obscene, indecent and profane language.

Literature Received

CATALOGUES

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Sub-miniature, miniature and standard lamps from Vitality Bulbs Ltd., 64 Marylebone Lane, London W.1, are described in catalogue no.69. WW 401 for further details.

Supplement no.2, dated February 1969, is available for the ITT (S.T.C.) Electronic Services Catalogue. ITT Electronic Services, Edinburgh Way, Harlow, Essex. WW 402 for further details.

Telemetry equipment for a wide range of applications and remote control systems are described in a brochure produced by Sound Diffusion Ltd., Datum Works, Hove, Sussex BN3 IRZ. WW 403 for further details.

Data on test and measuring equipment from B & K Laboratories Ltd., Cross Lances Rd., Hounslow, Middlesex, is given in a new short-form catalogue. WW 404 for further details.

Precision measuring apparatus (standard cells, potentiometers, galvos etc) are listed, complete with performance data, in a catalogue from H. Tinsley and Co. Ltd., Werndee Hall, South Norwood, London S.E.25. **WW 405** for further details.

An enlarged components catalogue, on the same lines as earlier editions, is now available from Home Radio (Components) Ltd., London Road, Mitcham, Surrey, price 8s 6d.

Redifon Ltd., of Broomhill Road, London S.W.18, have produced a catalogue which describes military and commercial communication equipment, broadcasting equipment and navigational aids. WW 406 for further details.

Catalogue of used scientific equipment from V. N. Barrett and Co. Ltd., 1 Mayo Road, Croydon, CRO 2QP, lists a wide range of items from vacuum pumps to electronic equipment. WW 407 for further details.

The 1969 **Transistor Catalog** of the Raytheon Company, Components Division, Semiconductor Operation, 350 Ellis Street, Mountain View, California, U.S.A., gives data on and physical details of a large number of transistor types. WW 408 for further details.

"Econoline", a moulded series of **plastic encapsulated transistors** from the Semiconductor Division of the Sprague Electric Company, Trident House, Station Rd., Hayes, Middlesex, are described in a short-form catalogue (CN200B). **WW 409** for further details.

High-power rectifiers (50kV p.i.v. at 1A) are listed in a catalogue from Solitron Devices Inc., 256 Oaktree Road, Tappan, N.Y. 10983, U.S.A. WW 410 for further details.

Croydon Precision Instrument Co. of Hampton Road, Croydon CR9 2RU, give details of a range of **precision components**, potentiometers, bridges, switches etc., in their latest catalogue. **WW 411** for further details.

APPLICATION NOTES

From Sprague Electric, Trident House, Station Rd., Hayes, Middlesex: TP.66.11 "Microcircuit digital to analogue converter", (WW 412 for further details); TP 68.24 "Series 5400/7400 integrated circuit application guide", (WW 413 for further details); and 25200 "More efficient logic design with multiple function series SE100 integrated circuits", (WW 414 for further details).

"Application and Characterization of a 250A Fast Recovery Rectifier" (AN-B-4) explains how the recovery time of a high-power rectifier is measured and how the construction of this rectifier (251UL) differs from normal rectifiers. Typical applications are also given. International Rectifier, 233 Kanas St., El Segundo, California 9024S, U.S.A. WW 415 for further details.

Digital-to-analogue converters for 4- to 10-bit words are described in application note 00011D/A from Sprague Electric (U.K.) Ltd., Trident House, Station Rd., Hayes, Middlesex. WW 416 for further details.

"Parameters ... Circuit Analysis and Design" is an 87-page application note (No.95) consisting of seven articles devoted to the description of high-frequency design (<100MHz) using "S" parameters. A description of "S" parameters is included. Hewlett Packard Ltd., 224 Bath Road, Slough, Bucks. WW 417 for further details.

Also from Hewlett Packard, of the above address, application note 920, "Harmonic Generation Using Step-recovery Diodes". WW 418 for further details.

PRODUCT LITERATURE

A dry-joint locator manufactured by Davian Instruments Ltd. is described in a leaflet available from Techmation Ltd., 58 Edgware Way, Edgware, Middlesex. WW 419 for further details.

Solderless wrapped "Barb" connectors are the subject of a leaflet from Oxley Developments Ltd., Priory Park, Ulverston, Lancs. WW 420 for further details.

Data on a 110-MHz digital frequency meter is given in a leaflet produced by Venner Electronics Ltd., Kingston By-pass, New Malden, Surrey. **WW 421** for further details.

A leaflet from Coutant Electronics Ltd., 3 Trafford Rd., Reading. Berks, gives details of a digital voltmeter (CDV200) with 0.05% accuracy and 100 μ V resolution. WW 422 for further details.

Performance details of a range of voltage reference valves (QS1200 to QS1213) and voltage stabilizers (OA, OB and QS series) are given in a brochure available from the English Electric Valve Co., Ltd., Chelmsford, Essex. WW 423 for further details.

Toroidal inductors from Control Technology Ltd., 44 Meeching Rd., Newhaven, Sussex, are the subject of a leaflet we have received. **WW 424** for further details.

"President" panel instruments, which are manufactured by Ferranti Ltd., Moston, Manchester M10 OBE, are described in a brochure that is now available. WW 425 for further details.

A y.i.g. tuned microwave receiver (1.8 to 25 GHz) designed for use, as a plug-in, with the Tektronix series 560 and letter-series oscilloscopes is described in a leaflet. The receiver can cover the whole band in one-sweep to display received signals on the tube face. Electro/Data Inc., 3121 Benton Street, Garland, Texas, U.S.A. WW 426 for further details.

For low-level measurements Tektronix have produced a 10 μ V/div. plug-in (Type 3A9) for the 560 series oscilloscopes which is described in a leaflet received. Tektronix U.K. Ltd., Beaverton House, Harpendon, Herts. WW 427 for further details.

GENERAL INFORMATION

"Equivalents Index 1969" lists the English Electric equivalents for a variety of valve types. English Electric Valve Co. Ltd., Chelmsford, Essex. WW 428 for further details.

"Mullard Data Book". The 1969 edition of this popular publication is now available at 3s 6d per copy from Mullard House, Torrington Place, London, W.C.1.

Details of a components brokerage service are available from GDS (Sales) Ltd., Michaelmas House, Salt Hill, Bath Rd., Slough, Bucks. **WW 429** for further details.

Details of the services offered by the National Research Development Corporation are given in a nicely produced brochure, called a "Service to Industry", we have received. N.R.D.C., 34 Bouverie St., London E.C.4. WW 430 for further details.

BS4410:1969, "Specification for the Connection of Flexible Cables and Cords for Appliances" is now available, price 6s, from the British Standards Institution, British Standards House, 2 Park St., London, W.1.

CP 1016: Part 1:1968 "The Use of Semiconductor Devices" is also available from the above address, price 12s.

Answers to **"Test Your** Knowledge''-12

Questions on page 244

1. (c). It is the particle displacement which is in phase quadrature with the excess pressure. A useful analogy can be drawn between excess pressure and particle velocity in a sound wave and electric and magnetic field strengths respectively in an electromagnetic wave. Thus we can define acoustic wave impedance as p/v and the instantaneous wave intensity can be shown to be p.v watts per square metre.

2. (b). The velocity is proportional to the square root of the abolute temperature. At frequencies above the audible range the velocity varies with frequency; at intensities well above the threshold of pain the velocity varies with intensity.

3. (d). This is an example of the Doppler effect and can give rise to a form of non-linear distortion in a cone loudspeaker which is simultaneously radiating a low and a high frequency tone.

4. (c).

5. (d). The musical interval between two notes depends on the ratio of their frequencies. Hence the frequency of F # must be $260 \times r$. Since $260 \times r \times r$ must = 520, $r = \sqrt{2}$.

6. (a). The definition of the loudness level of a sound in phons is the intensity, in dB above threshold, of a pure 1kHz reference tone which sounds equally loud.

7. (c). This is the approximate range between the threshold of audibility and the threshold of pain.

8. (a)

9. (c). For a normal moving coil loudspeaker the back e.m.f. (E) induced by the motion of the coil is B l v where v is the phasor velocity. The force on the coil is B 11 (1 being the phasor current in the coil) and this must equal $Z_m v$. Motional impedance = $E_{l} = (Bl)^2 / Z_{m}$.

10. (a). Provided the cone is fitted into a suitable baffle the acoustic power radiated does not change significantly as the frequency is increased from just above the fundamental mechanical resonance to at least a value giving a wavelength equal to the cone circumference. Below the mechanical resonance the power radiated falls off very rapidly with decreasing frequency.

11. (d). This effect is only significant at low frequencies

12. (c). The horn exhibits a cut off frequency which increases if the rate of taper with distance is increased. Below this frequency there is no appreciable sound transmission through the horn.

13. (b)

14. (d). At a pressure antinode in a standing wave we have a velocity node. The crystal microphone, being pressure operated, will give a large output; the ribbon microphone is a velocity microphone and therefore will give very little.

15. (a)

16. (c). A recommended range of reverberation times for rooms between 1000 and 3000 cubic foot volumes is 0.6 to 0.8 seconds above 1kHz rising to 1.2 to 1.6 seconds as the frequency drops to 30Hz.



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Ten pages of Electrical and Mechanical Data on our complete range of Moulded and laminated Switches. Brochure No. 1502/C.

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Real and Imaginary

by "Vector"

"In the day of adversity, consider"

Wireless World readers, as a race, are not given to writing letters to the journal unless some hapless author or the printer has mucked up an equation. Their reluctance is understandable, for in all probability most of their working week is spent writing reports which nobody reads. Such circumstances generate a certain disenchantment with penmanship to the point where the Christmas letter to Aunt Mary represents the total spare-time effort. I am therefore particularly grateful to the many readers who have been stirred to write to the editor or to me personally regarding my comments on take-overs and business structures in general. Many thanks for your most interesting letters, which have provided food for the thoughts which follow.

It does no harm at times to try to establish first principles. So let's think for a minute or so in general terms about the electronics industry; and in particular its importance in the scheme of things in the world today.

Of the last there can be no doubt. Electronics has so woven itself into the fabric of civilization that if all electronic devices simultaneously ceased to function the world would be in complete chaos. Communication services (both radio and line) would cease completely; aircraft would be grounded; manufacture and commerce disrupted—you name it, electronics plays a part in it. And, to complete the picture, try to imagine the demoralization of the public bereft of its sound radio set and its "goggle box"! *

This, then is something of the measure of the latent power of the electronics engineer. He has only to withdraw his services, his skills and his experience to overset the world.

I mention this because Professor D. A. Bell, in his thoughtful letter in the February issue, expresses doubts as to whether the learned societies could operate on the lines of the British Medical Association in protection of their engineers and technicians. You may remember, one point he made was that the electronics engineer does not carry the lifeand-death image which is attached to a doctor. This of course is very true. We don't carry much of an image at all with the public because, with the exception of the domestic radio servicing fraternity, we don't come into direct contact with them. Everyone knows, either at first or second hand, pretty well what goes on inside a hospital; nobody, when making a telephone call from

* Vector has not seen my mail!-ED.

London to Glasgow, thinks for a single moment of the complex electronic devices which come into operation when he does so. He doesn't even know they exist. But this doesn't mean that he wouldn't care if they ceased to function. He would care very much. Similarly with broadcasting. Nobody thinks of the engineers manning the studios and stations but if they withdrew their services the whole matter would be at parliamentary level within hours.

So much for the *potential* strength of the electronics engineering profession. In reality, however, it is pitiably weak. It is weak to the point of impotence because, as a generalization, we are individualists. We don't want to be dragooned by a militant trade union. We want to think our own thoughts, make our own decisions, take our own executive action. This is fine—provided that the premium doesn't rise too high.

In the post-war years, until the fairly recent past, the premiums were, on the whole, small. Immediately after peace broke out there was a famine of electronics engineers. We were lured, cajoled and cossetted by every blandishment known to the personnel recruitment boys; as Robert Browning *nearly* put it, "God was in His heaven, all was right with the world!" The firm for which we worked was smallish, but making a profit; moreover it was stable. Job-wise, the ambitious could confidently look forward to promotion; the others to security of tenure, provided the daily task was well done.

The one big mistake we made was to imagine that we had reached the millenium. The idea of building an electronic engineering profession defence mechanism, which had always been an unpalatable thought, was now ludicrous. No lab. stewards for us!

Then came the first rumblings of the storm. We read of small electronics firms being taken over by a big boy. So what? If there was a certain amount of redundancy, the unfortunate could easily get a job somewhere else. So we ignored the cloud on the horizon and concentrated on the clear blue sky overhead.

Now, with storms all around us, we are bewildered and aggrieved. The financial jugglers, with no interest in electronics other than the profits that can be wrung out of the industry, have moved in. Today there is no company or group, however large, which cannot be toppled into takeover. But even in adversity we are completely divided among ourselves. There is, for instance, the "up ladder, I'm aboard" school of thought adopted by some engineers in cases where the situation hasn't caught up with them and (they piously trust) it never will. Then there is the young, ambitious element who see in every enforced retirement a greater opportunity to further their own careers. And of course there are the redundants themselves, humiliated, bitter at being thrown prematurely on the scrap-heap, but alone and completely powerless to do anything about it. Lastly there is a core of engineers who take thought beyond mere self-preservation and the short-term future, but feel equally helpless to do anything constructive. It is upon the last-mentioned that the future status of the professional electronics engineer will depend.

Make no mistake about it, we are now paying an intolerably high premium for our individuality of outlook. We are discrete small sticks, easily snapped, whereas, if we had had the sense we were born with, we should have bound ourselves together for strength while the going was good. The financial jugglers know this and this is why the electronics industry is such an attractive fishing pond.

All right, then. Let's admit we were stupid. What to do about it?

If an opinion poll was taken I doubt whether many of us would opt in favour of a *trade* union. I can't readily see us coming out on strike because of a dispute as to which union man should be responsible for applying the solder to the bit. It still seems to me that the logical move would be to coerce the learned societies into throwing their formidable aggregate of weight in on our side. (I say "coerce" with intent because the institutions are not noticeably receptive to innovation. Official comment on the original suggestion was conspicuously absent.)

Clearly, there would be problems. A "best way" to go about it would have to be found-perhaps a committee to study the B.M.A. mechanism would be one promising approach. The question of costs would also have to be solved and here again the B.M.A. model seems to provide a basis for study. One particular headache which would arise concerns the large body of engineers and technicians who are not members of any learned society. One can also foresee that those who regard a learned society as an exclusive club would curl up at the thought of being associated with those who, for one reason or another, cannot put the magic letters after their names. Yet I know many such who possess neither university degree nor learned society membership, but who are nevertheless very able-and in one or two instances, distinguished engineers. What to do about them? Debar them and a singular injustice would be done. Admit them and you could be opening the door to all the quacks and incompetents.

No, it isn't going to be simple. But it can be done; indeed, to my mind, it must be done. From here on, the road forks three ways; towards a united front bonded together by the learned societies; towards a militant trade unionism; or to continue as we are, as extras in a twentieth century Uncle Tom's Cabin, with Simon Legree played by the Stock Exchange.

Think on, as they say up North.

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and 31 ips and mono/stereo. Frequency ±1dB. Rotary controls are used for volume, bass, treble and balance. Outputs may be fed to loudspeakers of any impedance from response extends from 15 Hz to 30 kHz 3 to 15 ohms. Case size-12 x 6 x 2 inches

station noise suppression, built-in AFC and ing discriminator for the best possible audio quality. Important features include interelectronic fine tuning, making the tuner a delight to handle. Tuning is from 86 to 108 MHz. A plug-in stereo decoder is available for adding to the mono version, or the tuner may be purchased complete in stereo form. THE VHF/FM TUNER uses a pulse counthigh (30 x 15 x 5 cm).

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Wireless World, May 1969

equipment for yourself, or write or phone for leaflet.

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Ri = 2KΩ, 20,000	100,000		
do 0-70°C 15,000)		
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Unmarked tra 2N753 1/6, B: LIGHT SEN ORP12 CAI 9/- each. GIANT-SIZE 6ma AT 0.6	SY28 SY28 SITIV DMIU	rs (tested) si 1/6, BSY65 E TRANS M SULPH LENIUM IS FROM	inilar t 1/6, O ITORS IIDE SOLA DAYL	o: C44 1/6, S (similar LIGHT-S R CELL IGHT!	OC71 1/-, to OCP71 SENSITIV	OC72 1/), 2/- each. YE RESIS	TORS P TC
NR 1351 14 Unmarked tra 20/753 1/6, B LIGHT SEN ORP12 CAI 9/- cach. GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD I 0.001µF 4000 0)0015µF 4000	SY28 SY28 SITIV DMIU SITIV DMIU SITIV SITIV POLY	rs (tested) si 1/6, BSY65 E TRANS M SULPH LENIUM TS FROM - each, 50m ESTER CA	milar t 1/6, O ITORS IIDE SOLA DAYL m. × 3d. 3d. 3d.	0: C44 1/6, S (similar LIGHT-S IGHT! 37mm. 2 1 FORS FA 0-15µF 0-22µF	OC71 1/-, to OCP71 EENSITIV .S—PROI for 10/ R BELOV 160V 160V	OC72 1/), 2/- each. E RESIS DUCE UI W COST PI 	TORS P TC RICE 6d 6d
NK 1351 4 Unmarked tra 2N753 1/6, Bi LIGHT SEN ORP12 CAI GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD 1 0.001µF 400 0.001µF 400 0.001µF 400 0.001µF 400 0.01µF 400 RECORD PI	SY28 SITTV SITTV DMIU SITTV DMIU SITTV DMIU SITTV DMIU SITTV	rs (tested) si 16, BSY65 E TRANS E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA	milar t 1/6, O ITORS IDE SOLA DAYL m, × PACI 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, 0 (similar LIGHT-S R CELL IGHT! 37mm. 2 TORS FA 0.15 µF 0.22 µF 0.27 µF 1 £F	OC71 1/-, to OCP71 EENSITIV .S—PROI for 10/ R BELOV 160V 160V 160V 125V	OC72 1/ 2/- each. E RESIS DUCE UI W COST PI 	TORS P TO RICE 6d 6d 1/-
NR 1351 1. Unmarked tra 21/753 1/6, B 21/753 1/6, B E 21/753 1/6, B E GIANT-SIZE 6 6ma AT 0.6 6 67mm, diamet 1 MULLARD 1 0 0001µF 4000 0001µF 4000 00012µF 400 0.0022µF 400 0.001µF 400 0.0022µF 400 0.001µF 400 0.0022µF 400 0.001µF 400 0.0022µF 400 0.001µF 400 0.0022µF 400 0.0022µF 400 0.0022µF 400	SY28 SITIV SMIU SE SEI VOLT ter 10/ VOLY VOLY VOLY VOLY VOLY VOLY VOLY VOLY	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA 	milar t 1/6, O ITORS IDE SOLA DAYL m. × PACI 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, 3 (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0.15µF 0.22µF 0.22µF 1£F 3. COMF Gibble £1, C	OC71 1/-, to OC771 SENSITIV SENSITIV for 10/ R BELO 160V 160V 160V 160V 125V PLETE W GP 93/1 C	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI 	FORS P TC RICE 6d 6d 6d 1/- DLES 0 25/-
NR 1351 14 Unmarked tra 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B GIANT-SIZE 6 6ma AT 0.6 67mm, diamet MULLARD I 0.001µF 400\ 0/001µF 400\ 0.0018µF 400 0.0018µF 400 0.0022µF 400 0.0022µF 400 0.0022µF 400 0.0022µF 400 CO18µF 400 0.0022µF 400 CO18µF 400 TRANSISTC KIT 10/-, CA VEP002041 VEP002041	SY28 SITIV DMIU SEI VOLT ter 10/ POLY/ V V V V LAYE no 15/- umic 2: DRISE	rs (tested) si 16, BSY65 E TRANS E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA R CARTRI S/ ED SIGNAI	milar t 1/6, O IITORS IIDE SOLA DAYL m. × PACI 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0-15µF 0-22µF 0-22µF 0:27µF 1£F S. COMI CTOR KI TT 10/-	OC71 1/-, to OC771 SENSITIV SENSITIV for 10/ R BELO V 160V 160V 160V 125V */LETE W GP 93/1 C	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI 	rors P TC RICE 6d 6d. 1/- DLES 0 25/- ACEF
NR 1351 4 Unmarked tra 20753 1/6, B LIGHT SEN ORP12 CAI GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD I 0.001µF 4000 0.001µF 4000 0.001µF 4000 0.002µF 400 0.002µF 400 0.002µF 400 0.002µF 400 0.002µF 400 0.002µF 400 0.002µF 400 0.002µF 400 TRANSISTO KIT 10/-, CA VEROBOAE 21 in × 1 in	SEI SY28 SITIV DMIU SEI VOLT ter 10/ POLY/ V V V V V V V V V V V V V V V V V V V	rs (tested) si 16, BSY65 E TRANS E TRANS F TRANS F TRANS F TRANS F TRANS F TRANS E STER CA CARTRI S CARTRI S	imilar t 1/6, O ITORS IIDE SOLA DAYL m. × PACI 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0-15µF 0-22µF 0-22µF 1£F S. COMI CTOR KI CTOR KI 11 10/-1 17 in. × a1	OC71 1/-, to OC71 1/-, SENSITIV SENSITIV for 10/ R BELO V 160V 160V 160V 125V PLETE W GP 93/1 C TT 10/-, SI 32 in. 0.	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI 	rors P TC RICE 6d 6d 6d 1/- DLES co 25/- ACEF
NR 1351 4 Unmarked tra 20753 1/6, B LIGHT SEN ORP12 CAL GLANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD I 0.001µF 4000 0.001µF 4000 0.001µF 4000 0.002µF 400 0.002µF 40	nsisto. SY28 STITV DMIU. 3 SEI VOL1 ter 10/ VOL1 ter 10/ ter	Ide, BSY65 TRANS T	milar t 1/6, O 1/70RS 1/10RS 1/10RS 1/10RS 1/10RS 1/10RS 1/10 1	0: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0-15µF 0-22µF 0-22µF 0:27µF 1£F S. COMI CTOR KI, CTOR KI, 17 in. × 31 in. × 32 in. × 31 in. ×	OC71 1/-, to OC71 1/-, S=NS1TIV SENSITIV for 10/ R BELO V 160V 160V 160V 125V PLETE W GP 93/1 C TT 10/-, SI 32 in. 0- 23 in. 0- 33 in. 0-	OC72 1/ 2 - cach. E RESIS DUCE UI W COST PI W COST PI W COST PI (TTH NEE] Irystal Stere IGNAL TR 1 matrix 1 matrix	TORS P TC RICE 6d 6d 6d 1/- DLES 0 25/- ACEF
NR 1351 1 Unmarked tra 1/6, B 21/753 1/6, B 21/753 1/6, B Suntain Sector 6/6 GIANT-SIZE 6 6ma AT 0.6 6 67mm, diamet 0.001µF 4000 0:001µF 4000 0:0018µF 400 0:0012µF 400 0:0022µF 400 0:0012µF 400 0:0022µF 400 0:0022µF 400 0:0022µF 400	nsisto. SY28 SITIV SITIV SITIV SITIV SITIV SITIV POLVIV V V V V V V V V V V V V V V V V V V	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA R CARTRI S/ ED SIGNAL V. COUNT matrix matrix matrix matrix	imilar ti 1/6, O. 11 1/6, O. 11 1/6, O. 11 1/6, O. 11 1/0 SOLA DAYL PACIT 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0-15µF 0-22µF 0-22µF 0-27µF 1£F 0:27µF 1£F CTOR KA CTOR KA 17 in. × 31 in. × 31 in. × 5 in. × 5 in. ×	OC71 1/-, to OC71 1/-, S=NS1TIV SENSITIV for 10/ R BELOV 160V 160V 160V 125V PLETE W GP 93/1 C TT 10/-, SI 32 in. 0- 32 in. 0- 33 in. 0- 33 in. 0-	OC72 1/ 2 - cach. 2 - cach. E RESIS DUCE UI W COST PI W COST PI W COST PI W COST PI 	TORS P TC RICE 6d 6d 6d 1/- DLES 0 25/- ACEF 14/- 4/- 4/- 4/- 4/- 5/-
NK 1351 1 Unmarked tra 2N753 1/6, B 1 UImmarked tra 2N753 1/6, B 1 URD TSEN 0 ORP12 CAI 0 GIANT-SIZE 6 6ma AT 0.6 6 67mm, diamet 0 MULLARD I 0 0001µF 4000 0001µF 4000 0:0022µF 400 0:0022µF 400	Insisto SY28 SITIV DAIU CONTENT SY28 SITIV DAIU POLYIV CONTENT SITING CONTENT SITIN SITING CONTI	rs (tested) si 16, BSY65 TRANS	milar t 1/6, O 1	0: C44 1/6, 3 (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0·15µF 0·22µF 0·22µF 1£F 3. COMH tible £1, C CTOR KI IT 10/ 17 in. × 5 33 in. × 5 5 in. × 5 10 C	OC71 1/-, to OC71 1/-, ENSITIV SENSITIV SENSITIV R BELOV 160V 160V 160V 125V PLETE W GP 93/1 C TT 10/-, SI 3 ³ in. 0 ⁻ 2 ¹ in. 0 ⁻ 3 ³ in. 0 ⁻ 2 ¹ in. 0 ⁻ 3 ³ in. 0 ⁻	OC72 1/) 2/- each. E RESIS DUCE UI W COST PI W COST PI W COST PI W COST PI W COST PI I antrix 1 matrix 1 matrix 1 matrix 2 matrix 1 matri	TORS P TC 6d 6d 6d 1/- DLES 0 25/- ACEF 14// 4// 4// 4// 5//
NR I 351 1 Unmarked tra 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B CRP12 CAI SCA ORP12 CAI SCA GRANT-SIZE 66ma AT 0.6 67mm, diameter MULLARD I MULLARD I 400 0.0018μF 4000 0.0022μF 400 0.0012μF 400 0.0012μF 400 0.0022μF 400 0.0022μF 400 0.0022μF 400	nsisto. SY28 S STIV DMIU, MU, STIV DMIU, STIV DMIU, VOLT ter 10/ VVOLT ter 10/ VVOLT t	rs (tested) si 1/6, BSY65 E TRANS M SULPH LENIUM TS FROM - each, 50m ESTER CA Constant 	milar t 1/6, 0 ITORS IIDE - SOLA DAYL PACI: 3d. 3d. 3d. 3d. 3d. 3d. 1DGES Compat Compat 1/3 3/31 5/6 11/- t Tool ter and	0: C44 1/6, 3 (similar LIGHT-S IGHT: TORS FA 0·15µF 0·22µF 0·22µF 1£F 0·27µF 1£F CTOR KI IT 10/ 17 in. × 31 in. × 5 in. × 5 in. × 9/6 Term 1 5 21 in	OC71 1/-, to OC71 1/-, S=NSITIV For 10/ R BELO 160V 160V 160V 125V PLETE W GP 93/1 C C C C C C C C C C C	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI 	TORS P TC 6d 6d 6d 6d 1/- DLES 0 25/- ACEF 14/- 4/- 4/- 5/-
NR 1351 4 Unmarked tra 20/753 1/6, B LIGHT SEN ORP12 CAI GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD I 0.001µF 4000 0.001µF 4000 0.001µF 4000 0.001µF 4000 0.002µF 400 0.002µF	nsisto. SY28 STIV DMIU, 3 SEI VOLT ter 10/ POLY1 VV VOLT ter 10/ VV VV VV VV VV VV VV VV VV VV VV VV VV	rs (tested) si 16, BSY65 TRANS	milar t 1/6, O ITORS IIDE : SOLA DAYL M. X : PACIT 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0-15µF 0-22µF 0-22µF 0-22µF 0:27µF 1£F S. COMI S. COMI IT 10/ 17 in. × 31 in. × 5 in. × 9/6 Term 185 0-0011 196 Conder	OC71 1/-, to OC71 1/-, S=NSITIV For 10/ R BELOV 160V 160V 160V 125V PLETE W GP 93/1 C PLETE W GP 93/1 C IT 10/-, SI 3 ² in. 0 ⁻ 3 ³ in. 0 ⁻ 3 ³ in. 0 ⁻ 3 ³ in. 0 ⁻ 3 ³ in. 0 ⁻ 3 ⁴ in. 0 ⁻ 5 ³ in. 0 ⁻ 5 ⁴ in. 0 ⁻ 5	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI W COST PI W COST PI ITH NEEI ITYSTAI Stere IGNAL TR I matrix I matrix I matrix I matrix I matrix J matr	TORS P TC 6d 6d 6d 6d 1/- DLES 50 25/- 14/- 4// 4// 4// 5// 0 only er 100 Mixeo
NK1351 1 Unmarked tra 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B GIANT-SIZE 6ma AT 0.6 67mm, diamet	nsisto. SY28 S STIV DMIU. 3 SEI TV DMIU. 3 SEI TV OL7 TV O	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM TS FROM - each, 50m ESTER CA - each, 50m ESTER CA CARTRI Soft ED SIGNAL V. COUNT matrix matrix matrix matrix matrix f Face Cut NSERS, M Ceramic, Pol /-, Pol 100. Sec 100. Sec 200. Sec 20	milar t 1/6, O ITORS IIDE I SOLA DAYL M. X 1 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0·15µF 0·22µF 0·27µF 1£F 0:27µF 1£F 0:27µF 1£F 0:27µF 1£F 0:27µF 1£F 0:15µF	OC71 1/-, to OC71 1/-, SENSITIV SENSITIV for 10/ R BELOV 160V 160V 160V 125V PLETE W GP 93/1 C PLETE W GP 93/1 C SI in. 0 ⁻ 3 ² in. 0 ⁻ 3 ³ in. 0 ⁻ 10 ¹ in. 0 ⁻ 5 ² i	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI 	P TC RICE 6d 6d 6d 1/- DLES 0 25/- 14// 4// 4// 4// 5// 0 only er 100 Mixed 5/-, pc
NK1351 4 Unmarked tra 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B GIANT-SIZE 6ma AT 0.6 67mm, diamet	nsisto. SY28 S STIV DMIU. 3 SEI VOL7 ter 10/ POLY1 VV OL7 ter 10/ POLY1 VV OL7 ter 10/ VOL7 ter 10/ POLY1 VV OL7 ter 10/ VOL7 ter 10/ SE ter 10/ SE ter 10/ SE ter 10/ SE ter 10/ SE ter 10/ SE ter 10/ SE ter 10/ SE ter 10/ SE ter 10/ SE SE ter 10/ SE SE ter 10/ SE SE ter 10/ SE SE ter 10/ SE SE ter 10/ SE SE ter 10/ SE SE SE ter 10/ SE SE SE SE SE SE SE SE SE SE SE SE SE	rs (tested) si 1/6, BSY65 E TRANS M SULPH LENIUM TS FROM - each, 50m ESTER CA COSTER CA CA COSTER CA COSTER CA 	milar t 1/6, O ITORS IIDE I SOLA DAYL M. X 1 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	0: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0:25µF 0:27µF 0:27µF 0:27µF 0:27µF 0:27µF 1£F 0:27µF 0:2	OC71 1/-, to OC71 1/-, SENSITIV For 10/ R BELOV 160V 160V 160V 125V PLETE W GP 93/1 C PLETE W GP 93/1 C PLETE W PLETE W PLETEPLETEPLEPLEPLEPLEPLEPLEPLEPL	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI 	TORS P TC RICE 6d 6d 6d 1/- DLES 0 25/- 14// 4// 4// 4// 5// 0 only er 100 Mixed 5/- pe or 10/-
NR 1351 1 Unmarked tra 21/753 1/6, B: 21/753 1/6, B: 21/753 1/6, B: 21/753 1/6, B: CRD17, SEN ORP12 CAL GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD I 0:001µF 400 0:001µF 400 0:001µF 400 0:001µF 400 0:001µF 400 0:001µF 400 0:002µF 400 0:001µF 400 0:001µF 400 0:002,µF 400 0:001µF 400 0:001µF 400 0:002,µF 400 0:002,µF 400 0:001µF 400 0:002,µF 400 0:0	Insiston SY28 STITU SMIU, MUL SMIU,	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA - cach, 50m ESTER CA Composition Composition Composition Composition Composition COUNT matrix matri	imilar t 1/6, O ITORS IIDE I SOLA DAYL M. X PACI: 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	o: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0·15µF 0·27µF 1£F 3. COMF CTOR KI IT 10/ 17 in. × 3 in. × 5 in. × 5 in. × 9/6 Term 1 6 22 in. × 1 6 27 in. × 5 in. × 16 2 in. × 5 in	OC71 1/-, to OC71 1/-, S=DS1TIV SENSITIV for 10/ R BELOV 160V 160V 160V 160V 125V PLETE W GP 93/1 C PLETE W GP 93/1 C TT 10/-, SI 32 in. 0 ⁻ 32 in. 0 ⁻ 33 in. 0 ⁻ 33 in. 0 ⁻ 33 in. 0 ⁻ 13 in. 0 ⁻ 15 in. 0 ⁻ 15 in. 0 ⁻ 15 in. 0 ⁻ 16 in. 0 ⁻ 17 in. 0 ⁻ 18 in. 0 ⁻	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI W COST PI W COST PI W COST PI W COST PI W COST PI W COST PI 	TORS P TC 6d 6d 6d 1/- DLES 0 25/- 14/- 4/- 4/- 4/- 4/- 5/- 0 only er 100 Mixee 5/- pe or 10/-
Unmarked tra 2N753 1/6, B: LIGHT SEN ORP12 CAL GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD 1 0-001µF 400 0-002µF 400 0-002µF 400 0-002µF 400 0-002µF 400 0-002µF 400 0-002µF 400 CRECORD PI GP 67/2 Mon GP 94/1 Cera TRANSISTC KIT 10/-, CA VEROBOAF Z4 in. × 24 in 5 pot Face Cu Special Offer PAPER CO SILVER-MIT types and value RESISTORS. 12 VOLT 7 NORMAL FOR 16 CH 10 CH 10 CH 10 CH 10 CH 10 CH 12 VOLT 7 NORMAL FOR 10 CH 10 CH 10 CH 10 CH 10 CH 10 CH 12 VOLT 7 NORMAL FOR 10 CH 10	Insiston SY28 STITU DMIU, 3 SEITU DMIU, 3 SEITU VOL1 VOL1 VOL1 VOL1 VOL1 VOL1 VOL1 VOL1	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM (S FROM - cach, 50m ESTER CA R CARTR , GP 91/3 (5/ ED SIGNAI MATIX matrix matrix matrix matrix matrix matrix matrix COUNT ESTER CA R CARTR , GP 91/3 (5/ ED SIGNAI COUNT MATRIX MATRIX MATRIX MATRIX SIGNAI SIGNAI SISTORIS 1, 9 watt 1 SISTORIS SISTORIS SISTORIS SISTORIS SISTORIS SISTORIS SISTORIS SISTORIS COUNT MPING OD	milar t 1/6, O ITORS IIDE - SOLA DAYL m. × PACI: 3d, 3d, 3d, 3d, 3d, 3d, 3d, 3d, 3d, 3d, 1DGES Compating (Compating) (Compatin	o: C44 1/6, (similar LIGHT-S IGHT! 37mm. 2 1 FORS FA 0.22µF 0.22µF 0.22µF 0.22µF 0.22µF 12F CTOR KI IT 10/ 17 in. × 31 in. × 5 in. × 5 in. × 9/6 Term 1 5 21 in ags 0.001µ te Conder cs, 2 to 1 0 10 watts y O.K. 7 LUORES ubc. Refi	OC71 1/-, to OC71 1/-, to OC71 1/-, S=NSITIV SENSITIV R BELOV 160V 160V 160V 125V PLETE W GP 93/1 C 125V PLETE W GP 93/1 C 100V 125V PLETE W GP 93/1 C 100V 125V PLETE W GP 93/1 C 100V 125V PLETE W GP 93/1 C 100V 100V 100V 100V 100V 100V 100V 100	OC72 1/), 2/- each. E RESIS DUCE UI W COST PI 	TORS P TC 6d 6d 6d 6d 6d 7/ 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0
NR 1351 4 Unmarked tra 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/753 1/6, B 21/754 GIANT-SIZE 6ma AT 0.6 67mm, diameter MULLARD I 0:001µF 4000 0:001µF 4000 0:001µF 4000 0:001µF 4000 0:002µF 4000 0:001µF 4000 0:002µF 4000 0:001µF 4000 0:001µF 4000 0:002µF 4000 0:001µF 4000 0:002µF 4000 0:001µF 4000 0:001µF 4000 0:002µF 4000 0:000 0:000µF 4000 0:000µF 4000 0:000µF 4000 0:000µF 4000 0:000µ	Insiston SY28 STITU SY28 STITU DMIU, MUL STITU VOLT VOLT VOLT VOLT VOLT VOLT VOLT VOLT	rs (tested) si 1/6, BSY65 E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA - cach, 50m ESTER CA R CARTR , GP 91/3 (5/ ED SIGNAI COUNT matrix COUNT SISTORIS SISTORIS CONDEN	milar t 1/6, O ITORS IIDE I SOLA DAYL M. X PACI: 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d.	o: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0·15µF 0·27µF 1£F 3. COMI Gible £1, C CTOR KI IT 10/ 17 in. × 3 in. × 5 in. × 5 in. × 9/6 Term 1 f 5 2 h in ags 0·001; te Conder es, } to 1 o 10 watts y O.K. 7 LUORES	OC71 1/-, to OC71 1/-, to OC71 1/-, ENSITIV SENSITIV R BELOV 160V 160V 160V 125V 21 ETE W GP 93/1 C 21 in. 0 23 in. 0 23 in. 0 24 in. 0 33 in. 0 24 in. 0 24 in. 0 34	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI W COST PI W COST PI W COST PI W COST PI W COST PI W COST PI W COST PI W COST PI W COST PI 	TORS P TC 6d 6d 6d 6d 1/- DLES 0 25/- 14// 4// 4// 4// 5// 0 only er 100 Mixed 5/- pe or 10/- HALLI 15 wat HGH7
NK 1351 1 Unmarked tra 2N/753 1/6, B: 2N/753 1/6, B: 2N/753 1/6, B: SU/753 1/6, B: GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD I 0:001µF 4000 0:002µF 400 0:001µF 400 0:001µF 400 0:001µF 400 0:001µF 400 0:002µF 400 0:001µF 400 0:001µF 400 0:002µF 400 0:001µF 400 0:002µF 400 0:001µF 400 0:002µF 400 0:002µF 30 YEROBOAR 2† in. × 1 in. 3† in. × 2† in 3† in. × 2† in 3† in. × 2† in 10: x 2† is Spot Face Cu Spot Face Cu Spot Face Cu Spot Face Cu NORMAL F0 1	nsisto. SY28 STIV DMIU. 3 SEIT VOL1 ter 10/ VOL1 ter 10/ ter 10/	rs (tested) si 1/6, BSY65 E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA - cach, 50m ESTER CA 	millar t 1/6, O ITORS IIDE 1 SOLA DAYL m. × 1 PACI: 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 1/3 3/11 3/11 5/6 11/- t Tool 1/2 in. t t maining ED F CUR SERS volt	o: C44 1/6, S (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0·25µF 0·22µF 0·27µF 1£F 3. COMI fible £1, C CTOR KI IT 10/ 17 in. × 5 33 in. × 5 34 in. × 5 9/6 Term 1 6 5 24 in ags 0·001; te Conder es, $\frac{1}{5}$ to 1 0 10 watts y O.K. 7; LUORES ubc. Refi RENT! 25µF 25µF 25µF 25µF 25µF	OC71 1/-, to OC71 1/-, te OC71 1/-, ENSITIV SENSITIV R BELOV 160V 160V 160V 125V 21ETE W GP 93/1 C 21 in. 0- 32 in. 0- 33 in. 0- 24 in. 0- 33 in. 0- 24 in. 0- 33 in. 0- 24 in. 0- 33 in. 0- 24 in. 0- 33 in. 0- 25 in. 0- 33 in. 0- 26 in. 0- 33 in. 0- 27 in. 0- 27 in. 0- 28 in. 0- 29 in. 0- 29 in. 0- 29 in. 0- 29 in. 0- 29 in. 0- 20 in.	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI · · · · · · · · · · · · · · ·	TORS P TC Gd Gd Gd Gd Gd Gd Gd Gd Gd Gd
NK 1351 1 Unmarked tra 2N753 1/6, B: LIGHT SEN ORP12 CAL 1 ORP12 CAL 6 GIANT-SIZE 6 6ma AT 0.6 6 67mm, diamet 0 MULLARD I 0 001µF 4000 0001µF 4000 0001µF 4000 0001µF 4000 0001µF 4000 00002µF 4000 0001µF 4000 0001µF 4000 00002µF 400 0001µF 4000 0001µF 400 00002µF 400 0001µF 400 0001µF 400 0001µF 400 00002µF 400 0001µF 400 00000 0002µF 400 0000 00002µF 400 00002µF TRANSISTCR 1 5 in. × 21 in 31 in 5 in. × 21 in 31 in 10 pot Face Cu Special Offer PAPER COO SILVER-MH types and yalt FOT <t< td=""><td>nsisto. SY28 S STIV DMIU. 3 SEI POLVIV VOLT ter 10/ VVOLT ter 10/ VVOLT ter 10/ VVOLT ter 10/ VOLT ter 10/ VO</td><td>rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA R CARTR: , GP 91/3 (5/ D SIGNAI W. COUNT matrix 18 watt 1 ESJ19/6, MPING OI 4µF 12 4µF 12 4µF 25</td><td>milar t 1/6, O ITORS IIDE : SOLA DAYL M. × PACI: 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 1DGES Compation SCOMPACI: 1/3 3/31 3/11 3/11 5/6 11/- t Tool 1/5 5/6 11/- t Tool 2 in t f CUR SERS volt volt volt</td><td>0: C44 1/6, 3 (similar LIGHT-S R CELL IGHT! 37mm. 2 FORS FA 0·25µF 0·22µF 0·22µF 0·27µF 1£F 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 15 µF 0·20µF 15 µF 0·20µF 15 µF 0·20µF 15 µF 0·20µF 15 µF 16 µF 15 µF 16 µF 16 µF 17 µF 17</td><td>OC71 1/-, to OC71 1/-, ENSITIV ENSITIV SENSITIV R BELOV 160V 160V 160V 125V 21 ETE W GP 93/1 C 21 in. 0- 21 in. 0- 2</td><td>OC72 1/ 2 - each. E RESIS DUCE UI W COST PI </td><td>TORS P TC 6d 6d 6d 6d 6d 7/ 0 0 0 0 0 0 0 0 0 0 0 0 0</td></t<>	nsisto. SY28 S STIV DMIU. 3 SEI POLVIV VOLT ter 10/ VVOLT ter 10/ VVOLT ter 10/ VVOLT ter 10/ VOLT ter 10/ VO	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM TS FROM - cach, 50m ESTER CA R CARTR: , GP 91/3 (5/ D SIGNAI W. COUNT matrix 18 watt 1 ESJ19/6, MPING OI 4µF 12 4µF 12 4µF 25	milar t 1/6, O ITORS IIDE : SOLA DAYL M. × PACI: 3d. 3d. 3d. 3d. 3d. 3d. 3d. 3d. 1DGES Compation SCOMPACI: 1/3 3/31 3/11 3/11 5/6 11/- t Tool 1/5 5/6 11/- t Tool 2 in t f CUR SERS volt volt volt	0: C44 1/6, 3 (similar LIGHT-S R CELL IGHT! 37mm. 2 FORS FA 0·25µF 0·22µF 0·22µF 0·27µF 1£F 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 15 µF 0·27µF 15 µF 0·27µF 15 µF 15 µF 0·20µF 15 µF 0·20µF 15 µF 0·20µF 15 µF 0·20µF 15 µF 16 µF 15 µF 16 µF 16 µF 17	OC71 1/-, to OC71 1/-, ENSITIV ENSITIV SENSITIV R BELOV 160V 160V 160V 125V 21 ETE W GP 93/1 C 21 in. 0- 21 in. 0- 2	OC72 1/ 2 - each. E RESIS DUCE UI W COST PI 	TORS P TC 6d 6d 6d 6d 6d 7/ 0 0 0 0 0 0 0 0 0 0 0 0 0
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NK 1351 1 Unmarked tra 21/753 1/6, B 21/753 1/6, B 1 SU/753 1/6, B 1 ORP12 CAL GIANT-SIZE 6ma AT 0.6 6 67mm, diamet 1 0.001µF 400 0001µF 400 0.001µF 400 0.001µF 400 0.001µF 400 0.002µF 400 0.002µF 400 0.002µF 400 0.001µF 400 0.002µF 400 0.001µF 400 0.002µF 400 0.002µF 400 0.002µF 400 0.001µF 400 0.002µF 400 0.002µF 400 0.002µF 400 IF 5 m. × 1 in. 3 in 3 in. × 2 in 5 in. × 2 in 5 in. × 2 in 1 in. 1	Insisto. SY28 STIL SY28 STIL DMIU, 3 SEL VOL1 ter 10/ POLY1 V VOL1 ter 10/ POLY1 V V V VOL1 ter 10/ POLY1 V V V VOL1 ter 10/ POLY1 V V V V V V V V V V V V V V V V V V V	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM CS FROM - cach, 50m ESTER CA R CARTR , GP 91/3 (5/ ED SIGNAI W. COUNT matrix SIGNAI V. COUNT SIGNAI V. COUNT SIGNAI SIGNAI SIGNAI CONDEN SISTORIS	milar triffer and the second s	0: C44 1/6, 3 (similar LIGHT-S R CELL IGHT! 37mm. 2 FORS FA 0·25µF 0·22µF 0·22µF 12F 0·27µF 12F 0:27µF 12F 0:27µF 12F 0:27µF 0:27µF 12F 0:27µF 0:27µF 0:27µF 0:27µF 0:27µF 0:27µF 0:27µF 0:27µF 0:27µF 0:27µF 12F 0:27µF 15 In. × 5 5 in. × 5 in. × 16 CTOR KI 17 10/ 17	OC71 1/-, to OC71 1/-, te OC71 1/-, ENSITIV SENSITIV R BELOV 160V 160V 160V 125V PLETE W GP 93/1 C C 125V PLETE W GP 93/1 C C 125V C 12	0C72 1/ 2 - each. E RESIS DUCE UI W COST PI · · · · · · · · · · · · · · · · · · ·	TORS P TC 6d 6d 6d 6d 6d 6d 6d 6d 6d 6d
NK 1351 4 Unmarked tra 2N/53 1/6, B: LIGHT SEN ORP12 CAI GIANT-SIZE 6ma AT 0.6 67mm, diamet MULLARD 1 0.001µF 400 0.001µF 400 0.001µF 400 0.001µF 400 0.002µF 4	Insistor SY28 STITU SY28 STITU DMIU, 3 SELT VOLT ter 10/ POLVIV VOLT ter 10/ POLVIV TE	rs (tested) si 16, BSY65 E TRANS M SULPH LENIUM CS FROM - cach, 50m ESTER CA R CARTR: , GP 91/3 (5/ ED SIGNAI W. COUNT matrix SISTORIS	millar t 1/6, O ITORS IIDE : SOLA DAYL M. × PACT: 3d. 3d. 3d. 3d. 3d. 3d. 3d. 1DGES Compatible SCOMPACT: 5/6 11/- t Tool 1/3 3/11 3/11 5/6 11/- t Tool 2 in, t R SERS Volt volt volt volt volt volt volt volt v	0: C44 1/6, 3 (similar LIGHT-S R CELL IGHT! 37mm. 2 1 FORS FA 0·25µF 0·22µF 0·22µF 1£F 0·27µF 1£F 33, in. × 5 in. × 9/6 Term 15 24 in 15 24 in ags 0·001; e CONGES ube. Refi 25µF 25µF 25µF 25µF 25µF 25µF 50µF 50µF 50µF 50µF 50µF 50µF 50µF 50µF 50µF	OC71 1/-, to OC71 1/-, te OC71 1/-, te OC71 for 10/ R BELO 160V 160V 160V 125V PLETE W P 93/1 C C T 10/-, SI 32 in. 0· 32 in. 0· 32 in. 0· 33 in. 0· 33 in. 0· 34 in. 0· 34 in. 0· 34 in. 0· 35 in. 0· 35 in. 0· 35 in. 0· 35 in. 0· 36 in. 0· 37 in. 0· 37 in. 0· 37 in. 0· 38 in. 0· 38 in. 0· 38 in. 0· 38 in. 0· 39 in. 0· 39 in. 0· 50 in	0C72 1/ 2 - each. E RESIS DUCE UI W COST PI 	TORS P TO RICE: 6d 6d 6d 6d 71/- DLES 0 25/- DLES 0 25/- 14:/ 4/2 4/2 4/2 4/2 4/2 4/2 4/2 4/
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AR8 5/- ARP3 3/- ARP12 3/6 ARP12 3/6 ARP12 3/6 ATP4 4/3 AZ21 9/6 BD78 40/- BL3 10/- BT35 16/- BT45 150/- BT83 35/- CV102 3/- CV102 3/- CV103 4/- CV315 (matched	ECC82 4/3 ECC83 5/6 ECC84 6/- ECC84 6/- ECC86 7/- ECC86 7/- ECC89 4/- ECC89 6/6 ECC89 6/6 ECC83 11/- ECC81 210/- ECC81 20/- ECC83 8/6	G/371K G1/370K 20/- G50/2G 20/- G120/1B G120/1B GU- GX23 9/6 GZ34 H141 H1/23DD H141 KT86 18/6 KT66 18/7 KT77 7/6	PY83 6/6 PY83 7/- PY800 9/- QV03-10 9/- QQV03-10 19/- QQV06-40 85/- QS150-18 -/ QS150-18 -/ QS150-18 -/ QS1202 8/- QV047 8/- QZ04-15 57/6 R10 17/6 R17 8/6 R19 7/6	U25 14/6 U26 14/6 U27 8/- U51 4/6 U53 4/6 U54 4/6 U54 4/6 U50 12/6 U50 12/6	VR150/30 6/- VU39 7/- V118 8/- W119 8/- X65 6/- X66 7/6 X118 8/- X66 7/6 X118 8/- X66 7/6 X118 8/- Z6007 28/- Z8007 28/- Z8007 28/- L03GT 6/- L03GT 6/- L4 2/6 L4 2/6 L4 2/6 L4 3/-	3Q4 8/- 3Q3GT 6/- 3R4 5/9 3V4 6/6 4D1 4/- 5D351 5/- 5D351 5/- 5D351 5/- 5D351 5/- 5D351 5/- 5D354 5/- 5D354 5/- 5D364 5/- 5D364 5/- 5D364 5/- 5C40 7/- 5V307 5/- 5Y3WGB 5Y3WGTB 5Y3WGTB 9/-	6 A M6 3/- 6 A N5 20/- 6 A N5 20/- 6 A Q5 5/6 6 A Q5 W 9/- 6 A 86 6/- 6 A 76 4/6 6 A 76 4/6 6 A 76 4/6 6 B 4/7 12/8 6 B 4/6 6 B 1/7 7/- 6 B 1/6 7 6 B 1/6 7 A (- 6 B 7 9/-	6EA8 9/- 6EU7 7/- 6F637 8/- 6F637 8/- 6F76 8/- 6F7 8/- 6F7 8/- 6F7 8/- 6F7 8/- 6F7 8/- 6F12 4/- 6F13 5/- 6F33 3/- 6F33 3/- 6F33 3/- 6G607 2/6 6H61M 19- 6J6407 7/- 6J507 5/- 6J6 3/6 6J760 5/- 6J760 5/- 6J70 5/- <tr td=""> <tr td=""></tr></tr>	68J7 5/- 68J70 6/6 68J70 7/- 68L70 7/- 702 1/- 702 1/- 702 1/- 704 1/- 707 6/- 707 7/- 704 1/- 707 6/- 707 7/- 707 7/- 70	COUCH THE VALV GUAR. 28D7 6/- 30C15 15/- 30C15 15/- 30C18 15/- 30C	ANTEE 957 5/- 9584 4/- 962 6/- 1625 6/6 1625 6/6 1
pairs) 120/- CV315	ECL82 6/- ECL83 10/9	KT88 27/-		TRANSI	STORS, ZI	NER DIO	DES, ETC.		707 7/-	30PL1 14/-	5704 9/-
CV315 (single) 50/- CV31 7/6 D41 3/- D41 3/- D41 3/- D4790 7/6 D4790 7/6 D4790 7/6 D4790 7/6 D4790 2/- D4790 2/- D490 7/6 D490 7/6 D490 2/- D490 2/	$\begin{array}{c} {\rm ECL483} \ 10/9\\ {\rm ECL483} \ 3/6\\ {\rm EP36} \ 3/6\\ {\rm EP36} \ 3/6\\ {\rm EP37} \ 8/-\\ {\rm EF39} \ 8/-\\ {\rm EF39} \ 8/-\\ {\rm EF39} \ 8/-\\ {\rm EF42} \ 13/6\\ {\rm EF49} \ 3/-\\ {\rm EF90} \ 4/6\\ {\rm EF96} \ 6/3\\ {\rm EF96} \ 8/-\\ {\rm EF99} \ 3/-\\ {\rm EF91} \ 3/-\\ {\rm EL35} \ 5/-\\ {\rm EL34} \ 10/3\\ {\rm EL41} \ 9/9\\ {\rm EL35} \ 5/-\\ {\rm EL31} \ 10/3\\ {\rm EL42} \ 10/3\\ {\rm EL42} \ 3/-\\ {\rm EM31} \ 3/-\\ {\rm EM30} \ 7/-\\ {\rm EM30} \ 7/-\\ {\rm EM30} \ 7/-\\ {\rm EM30} \ 7/-\\ {\rm EM31} \ 3/-\\ {\rm EM31} \ 3/-\\ {\rm EM31} \ 3/-\\ {\rm EM31} \ 8/-\\ {\rm EM31} \ 8/-$	RT243 7/- L43 6/- MH41 8/- OB3 7/- OB3 7/- OC3 6/- OB3 7/- PABC80 7/- PABC80 7/- PC024 8/- PC03 6/- OC3 6/- PC84 6/- PC0480 10/3 PC0740 8/- PC084 6/3 PC084 6/3 PC180 10/9 PC181 9/- PC184 8/3 PC184 8/3 PC184 8/4 PC184 8/4 PC184 8/- PC184 8/- PC184 8/- PC184 <td< td=""><td>0A5 3/- 0A7 4/- 0A10 3/- 0A10 8/6 0A10 8/6 0A70 8/6 0A70 8/6 0A71 8/6 0A73 8/6 0A74 8/- 0A75 2/6 0A74 8/- 0A75 2/6 0A79 2/6 0A80 8/- 0A90 2/- 0A202 3/6 0A22010/- 0A22016 0A22021 to 0A22021 to 0A22021 to 0A222016 0A222021 to 0A22202 to 0A222021 to 0A22213 6/6 0A22203 to 10/A2223 10/- 0C223 10/- 0C223 10/6 0C223 10/6 60/- 0C223 10/6 60/- 0C223 10/- 0C23 10/6 0C23 10/6 3/3 BTV280/40 80/- STV280/80 80/- SU2150A 10/-<</td><td>0 C38 13/6 0 C39 13/6 0 C39 13/6 0 C39 13/6 0 C38 12/6 0 C38 12/6 0 C38 12/6 0 C34 13/- 0 C44 5/- 0 C44 5/- 0 C44 5/- 0 C73 6/- 0 C31 3/- 0 C32 5/- 0 C31 3/- 0 C33 5/- 0 C122 10/- 0 C133 7/6 0 C133 7/6 0 C133 7/6 0 C133 5/- U141 9/9 UL41 9/9 UL41 9/9 UL41 9/9 UL41 9/9 UL41 9/9 UL41 9/9 UL41 3/- 0 C13 3/- V Y133 3/- V Y133 3/- V Y130 3/- V R99 7/6 V R105/30 6/-</td><td>OC170 7/8 OC270 7/6 OC200 1/6 OC201 10/8 OC201 10/8 OC202 10/8 OC202 10/8 OC202 10/8 OC204 17/6 OC205 17/6 201001 19/6 201303 10/7 201303 10/7 201303 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ECC40 10/9 ECC81 4/-	FW4/800	PY81 5/6 PY82 5/9	U12/14 7/- U18 6/-	free. C.0	D. D. 4/- extra.	£1 2/- Ln £, over	r £3 post	68C7GT 5/-	25Z5 8/6	955 2/6	725A
NAGARD C DE 103, £85. PORTABLE LOSCOPE 2 £22 108. Carr HEWLETT 524B ELECT Without plug will measure f 10-1 mc/s and Frequencies a decimal point	SCILLOSCO Carriage 10/ SONTRANIO 13 age 30/ -PACKARD TRONIC CO in unit this i requencies from periods of from re read in kc/s automatically p	PE TYPE C COSCIL- 250v. A.C., 2 MODEL 7 DUNTER. a nostrument 2 n 10 c/s to 2 s with the s with the	0.C. MOVING 00mA 21 in. squ 0-20mA 2 1n. Ro 0-0-30mA 21 in. Ro 0-0-30mA 21 in. Ro 0-0-30mA 21 in. squa nd figures	COIL MET are panel ound panel re, black dial lu panel d panel d panel d panel	ERS minous hand d calibro-30	book an Carriage 22/6 Carriage 20/- SPARE: 17/6 Ask for 20/- selection 18SET 12/6 FIELD 22/6 FIELD 22/6 Housed 22/6 Excellen 30/- out-door 30/- out-door	d necessary c 30/ SFOR AR.88E your needs MICROPHC andset, 2/6. P. TELEPHON in portable t for commun rs for up to 10 batteries and	onnector. £45. A RECEIVERS. from our huge ONE for tele- & P. 2/ ES TYPE "F": wooden cases. ication in- and miles. For pair //6th mile field	SMALL 28 approx. 4,0 fans, runnin grinders, et MECHANI RELAYS. working fro delay within P. & P. 3/ HIGH SPE PLUG IN	V MOTORS. 00 r.p.m. Ide g models, min c. 12/ P. & ICAL TIME! Coil resistanc: m 12-40 D.C. range of few st ED ULTRA S I RELAYS	150/200mA al for small iature drills, P. 2/ D DELAY e 150 ohms, Adjustable sconds. 17/
or microsecon	d in seconds, m ds again with t	he decimal 5	mA. 2in. round	clip-fix panel of	or proj	20/- cable of	n drum. Co	Mpletely new,	separate wir	dings each of	1685 ohms.
tion is in eight	t places, first si	x on neon	0-0-10 mA. 21in.	round panel		17/6 10/	TELEBLIA	TYPE HIS	LOW INE	TIA 24V D.C	MOTOR,
lamp decades,	last two on me	eters. Self 7	5 mA. 21in. plug	In		14/- As above	e but in porta	ble metal cases.	complete wi	th gears. 15/	P. & P. 3/
10 mc/s freque	ncy standards.	Full details	00 mA. Itin. pro	j	*****	17/6 Per pair	d cable on dru	eries and 1/6th m. Slightly used	SHUNTS	25/-, P. 8	P. 3/
extra range, IC	$\frac{100}{220}$ mc/s, is a	n optional li	00 mA. 21in. rou	ind panel		19/- but gu	aranteed wor	king. £5/10/			
EXTRA £22/10/	- Carriage 15	VALVE 2	00 mA, 21 in. ro	und panel		17/6 HARNI	ESS "A" & "B	control units,	FOR	EXPORT O	NLY
VOLTMETE	RTYPE 378 B/2	Accurate 2	5 amp. 3 in. rou	nd proj		27/6 phones.	boxes, head: etc.	phones, micro-	Installation	Kits for CI	RZIO Sets
250 kc/s in the	ranges 10mV	(full scale) 50	amp. 2 in. rou	nd panel	ound mand	27/6 29/4IFT.	AERIALS ea	ch consisting of	new" stand	ard. All spar	es available.
divided. A db	scale). Logarit scale provideo	for 0-20 20	0 VDC 2in. squa	re panel	ound panel	19/- sections.	Ilft. (6-section	on) whip aerial	COLLINS	TCS. Compl	ete installa-
db, 0 db being	g ImV. Automa	atically set 10	00 V 4in. round	panel		15/- with ada	ptor to fit the	7in. rod, insu-	tions and spa	re parts.	
for monitorir	ig the input	signal if	50-0-1500 mA. 3	in. round pan	el	25/- pegs, rea	mer, hammer,	etc. Absolutely	C42 & C45.	12v and 24v.	ITS FOR
Post and pack	ing 10/	LL/10/ 1.	5 KV with res. 2	lin. round pan	el	27/6 in canvas	bag, £3/9/6. P	. & P. 10/6.	RECEIVERS	S R 210.	
RF WATT M	ETER PMI6.	Frequency	E METERS			300W	set) 635 Ca	trol Generator	R.C.A. TRA 4336. 2-20 M	ANSMITTER 1c/s., complete	with M.O.
0-1,500w. Imp	edance 51.5 of	hms. "N"	20 mA 21 in co	und panel		1260W 3	SV CHARGI	NG SET. Com-	Cryst. mul	t. and spe	ech ampl.
MINIATURE	E METERS.	General 4	amp. 2 in. round	d panel		22/6 Carriage	40/	a. New 65.	available.	and gueranteed	• All spares

 PHINATORE METERS. General values, clip nounded gamet in round panel in round panel in round gamet in round panel in round pan







BRAND NEW and in original cases—A.C. mains input. 110V or 250V. Freq. in 6 bands 535 Kc/s-32 Mc/s. Output impedance 2.5-600 ohms. Complete with crystal filter, noise limiter, B.F.O., H.F. tone control, R.F. & A.F. variable controls. Price £87/10/each, carr. £2.

Same model as above in second hand cond. (guaranteed working order), from $\pounds 45$ to $\pounds 60$, carr. $\pounds 2$.

*SET OF VALVES: new, £3/10/- a set, post 7/6; SPEAKERS: new, £3 each, post 10/-. *HEADPHONES: new, £1/5/- a pair, 600 ohms impedance. Post 5/-.

AR88 SPARES. Antenna Coils L5 and 6 and L7 and 8. Oscillator coil L55. Price 10/- each, post 2/6. RF Coils 13 & 14; 17 & 18; 23 & 24; and 27 and 28. Price 12/6 each. 2/6 post. By-pass Capacitor K.98034-1, 3×0.05 mfd. and M.980344, 3×0.1 mfd., 3 for 10/-, post 2/6. Trimmers 95534-502, 2-20 p.f. Box of 3, 10/-, post 2/6. Block Condenser, 3×4 mfd., 600 v., **£2** each, 4/- post. Output transformers 901666-501 **27**/6 each, 4/- post.

· Available with Receiver only.

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MARCONI SIGNAL GENERATORS

TYPE TF-144G

Freq. 85Kc/s-25Mc/s in 8 ranges. Incremental: +/-1% at 1Mc/s. Output: continuously variable 1 microvolt to 1 volt. Output Impedance: 1 microvolt to 100 millivolts, 10 ohms 100mV-1 volt-52.5 ohms. Internal Modulation: 400 c/s sinewave 75% depth. External Modulation: Direct or via internal amplifier. A.C. mains 200/250V, 40-100 c/s. Consumption approx. 40 watts. Measurements: $19\frac{1}{4} \times 12\frac{1}{4} \times 10$ in. The above come complete with Mains Leads, Dummy Aerial with screened lead, and plugs. As New, in Manufacturer's cases, £40 each. Carr. 30/-. DISCOUNT OF 10% FOR SCHOOLS, TECHNICAL COLLEGES, etc.

3-B TRULOCK ROAD, TOTTENHAM, N.17

Phone: Tottenham 9213

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ALL GOODS OFFERED WHILST STOCKS LAST IN	"AS IS" CONDITION UNLESS OTHERWISE STATED
DE-ICER CONTROLLER MK. III: Contains 10 relays D.P. changeover heavy duty contacts, 1 relay 4P, C/O. (235 ohms coil). Stud switch 30-way relay operated, one five-way ditto, D.C. timing motor with Chronometric governor $20-30$ v., 12 r.p.m.; geared to two 30-way stud switches and two Ledex solenoids, 1 delay relay etc., sealed in steel case (4 × 5 × 7 ins.) \$3 each, post 7/6.	PRD Electronic Inc. Equipment: STANDING WAVE DETECTOR: Type 219, 100-1,000 Mc/s. (New) £65 each, post 12/6. FREQUENCY METER: Type 587-A, 0.250-1.0 KMC/SEC. (New) £75 each, post 12/6. FIXED ATTENUATOR: Type 130c, 2.0-10.0 KMC/SEC. (New) £5 each, post 4/-, FIXED ATTENUATOR: Type 1157S-1, (new) £6 each, post 5/-,
TX DRIVER UNIT: Freq. 100-156 Mc/s. Valves 3 × 3C24's; complete with filament transformer 230 v. A.C. Mounted in 19in. panel, £4/10/- cach, 15/- carr. POWER UNIT : 110 v. or 230 v. input switched; 28 v. @ 45 amps. D.C. output. Wt. approx. 100 lbs., £17/10/- each, 30/- carr. SMOOTHING UNITS suitable for above 6/10/e cach, 15/e carr.	TERMALINE RESISTOR UNITS: type 82A/U, 5000W, freq. 0-3.3 KMC Max VSWR 1.2 Type "N" female connectors, etc. Brand new, £30 each, carr. 15/
POWER SUPPLY UNIT PN-12B: 230 v. A.C. input, 395-0-395 v. output @ 300 mA. Complete with two × 9H chokes and 10 mfd. oil filled capacitors. Mounted in 19in. panel, £6/10/- each, £1 carr.	COAXIAL IEST EQUIPMENT: COAXWITCH—Mnftrs. Bird Electronic Corp. Model 72RS; two-circuit reversing switch, 75 ohms, type "N" female connectors fitted to receive UG-21/U series plugs. New in ctrs., £6/10/- each, post 7/6. CO-AXIAL SWITCH—Mnftrs. Transco Products Inc., Type M1460-22, 2 pole, 2 throw. (New) £6/10/- each, 4/6 post. 1 pole, 4 throw, Type M1460-4 (New) £6/10/ each 4/6 post. 1 pole, 4 throw,
OHMITE VARIABLE RESISTOR: 5 ohms, 5 ¹ / ₂ amps; or 2.6 ohms at 4 amps.	
CONTROL PANEL: 230 v. A.C., 24 v. D.C. @ 2 amps., £2/10/- each, carr. 12/6. AUTO TRANSFORMER: 230-115 v.; 1,000 w. £5 each, carr. 12/6. 230-115 v.; 2001/4 .52 each	DECADE RESISTOR SWITCH: 0.1 ohm per step. 10 positions. 3 Gang, each 0.9 ohms. Tolerance $\pm 1\%$ £3 each, 5/- post. 90 ohms per step. 10 positions, total value 900 ohms. 3 Gang. Tolerance $\pm 1\%$ £3/10/- each, 5/- post.
SOLENOID UNIT: 230 v. A.C. input, 2 pole, 15 amp contacts, £2/10/- each post 6/	post. Excellent secondhand cond. (Meters only—batteries and leads extra, at cost.)
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38-hour, 5-day week with usual holiday and slck pay schemes. Opportunities for evening work with additional pay. Permanent pensionable posts; medical.

Further particulars and application forms (to be returned within fourteen days) from Registrar, Bristol Technical College, Ashley Down, Bristol, BS7 9BU. Please quote appropriate reference

number in all communications.

MEL MANAGEMENT SELECTION

Please state briefly, in writing, how each requirement is met. Nothing will be disclosed, unless you give permission after a confidential interview.

HEPUIK

Advertisements accepted up to MAY 9 for the JUNE issue, subject to space being available.

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VIENIS

Development Engineer

radio telecommunications (Ireland)

This Irish company, employing about 200 but part of a major international electrical and electronic group, specialises in the design, development and quantity production of radio telecommunications and specialised peripheral test equipment, much of it unique.

The company seeks two radio communications development engineers who must have had experience in taking a project from initial specification through the design and development stages to final quantity production. They must be professionally qualified, either graduates or AMIERE and should have had some actual quantity production experience. Ideal age about 30. Salary £2000. Location Dublin. Please send details of age, qualifications and experience to P. J. H. Fryer quoting ZH.80552.

MSL IRELAND LTD. 49 Upper Mount Street, Dublin 2

C.S.E. (AIRCRAFT SERVICES) LTD. Oxford Airport • Kidlington

Electrician Electronic Wireman Radio Technician

It you are one of these, a job awaits you at C.S.E. (Aircraft Services), Kidlington. The work covers the preparation of Wiring and Components and the Installation into modern aircraft (including the Jetstream), of Radio and Electronic Systems Equipment.

Weekly Staff Status, Pension Scheme, Free Life Assurance, Sickness Allowance.

Apply in first instance to :-

THE ADMINISTRATION SUPERVISOR C.S.E. (AIRCRAFT SERVICES) LTD. OXFORD AIRPORT KIDLINGTON OXFORD Telephone: Kidlington 3931 **GEC-Marconi Electronics**

ELECTRONIC TEST ENGINEERS

Our Test Department is responsible for testing and fault finding on a wide range of Marconi equipment; airborne communication and navigation aids; radar; broadcasting; and space, radio and line communications. There are excellent career prospects both within the Test Department and in other areas of the expanding Marconi Company.

We wish to hear from men with a proven career record in the electronics industry who, preferably, should have gained qualifications to at least C & G Telecommunications Intermediate standard.

Members of H.M Forces in the electronic fitter category would find these positions of particular interest.

Please write for brochure to Mr. M. J. Shepherd, Personnel Officer, Chelmsford Works, The Marconi Company Limited, Marconi House, Chelmsford, Essex, quoting reference WW/E/61.

Member of GEC-Marconi Electronics Limited

APPOINTMENTS

GOOD SALARY AND CONDITIONS.

EXISTING HOLIDAY ARRANGEMENTS HONOURED.

Please send full details in writing to:-MANAGING DIRECTOR,

HAYDEN LABORATORIES LTD., EAST HOUSE, CHILTERN AVE., AMERSHAM, BUCKINGHAMSHIRE.

UNIVERSITY OF SOUTHAMPTON

Department of Electronics

An EXPERIMENTAL OFFICER experienced in digital techniques required to take charge of the day-to-day running of several research and teaching projects. A number of the projects are linked to a Honeywell 516 computer and familiarity with a computer or similar system is essential. Applicants should be graduates or hold associate membership of a relevant professional institution.

A JUNIOR TECHNICIAN is required to work in the field of microelectronics. Some knowledge of chemistry or photography is desirable for this post, but full training will be given. Four G.C.E. 'O' levels required.

Salary scale for Experimental Officers rises to £1,930 per annum plus F.S.S.U.

Salary scale for Junior Technicians £352-£595 according to age and qualifications.

Applications, giving details of age, qualifications, experience and the names of two referees, should be sent to the Deputy Secretary, The University, Southampton, SO9 5NH, guoting reference WW.

Government of ZAMBIA REQUIRES

RADAR ENGINEER

for the Department of Meteorology, Ministry of Transport, Power and Works, on contract for one tour of 36 months in Power and Works, on contract for one four of 36 months in the first instance. Commencing salary according to experience in scale Kwacha 2,736 (£Stg. 1,596) rising to Kwacha 3,216 (£Stg. 1,876) a year, plus an inducement allowance of £Stg. $568 - \pounds$ Stg. 615. Gratuity 25% of total salary drawn. A direct payment of £Stg. 268 - £Stg. 291 is also payable direct to an officer's U.K. bank account. Both gratuity and direct pay-ment are normally TAX FREE. Free passages. Quarters at low rental Children's education allowances Liberal leave on low rental. Children's education allowances. Liberal leave on full salary or terminal payment in lieu. Contributory Pension Scheme available in certain circumstances.

Candidates between 22 and 35 years of age, must

- 1. Have served a five years apprenticeship in radio and radar engineering, or
- 2.
- Possess a service Trade Certificate, or Possess a City and Guilds Intermediate Certificate in Telecommunications Engineering or its equivalent.

Preference will be given to candidates experienced with

- 1. H.F. R/T transceivers radio-facsimile and radio-sonde.
- 2. S and X-band radar equipment.

Duties include :-

a. Repair and maintenance of all radar sets and communications equipment for which he is responsible, plus the

b. Care and Maintenance of appropriate spares and stores.

He may also be required to assist in installation work.

Write to CROWN AGENTS 'M' Dept., 4 Millbank, London, S.W.1, for application form and further particulars, stating name, age, brief details of qualifi-cations and experience and quoting reference M2K/ 690222/WF

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Careers in

ELECTRONICS with the

AIR FORCE DEPARTMENT

Vacancies at RAF Sealand, near Chester RAF Henlow, Bedfordshire and RAF Carlisle, Cumberland

Interesting and vital work on RAF radar and radio equipment for:

TELECOMMUNICATIONS TECHNICAL OFFICERS GRADE 3

Minimum qualification, ONC in electrical engineering or equivalent qualification.

Starting pay according to age, up to £1347 p.a. (at age 28) rising to £1521 p.a. with prospects of advancement to higher grades with pay scales up to £2.500 p.a.

RADIO TECHNICIANS

Minimum qualification, 3 years' training and practical experience in radio engineering.

Starting pay according to age, up to £1130 p.a. (at age 25) rising to £1304 p.a. with prospects of promotion to T.T.O. Grade 3 (above).

5 day week—good holidays—help with further studies—opportunities for pensionable employment.

Write for further details to: Ministry of Defence, CE3h(Air), Sentinel House, Southampton Row, LONDON, W.C.1, stating post required.

Applicants must be U.K. residents.

A senior man is required to take charge of the Drawing Office of an expanding wholly British privately owned company engaged in the invention, design and manufacture of electronic components.

The modern factory, ideally situated in the Lake District, employs a very wide variety of talents and offers unequalled opportunities to a competent, educated, inventive and level headed man, preferably with a wide peripheral knowledge of physics and chemistry

The usual advantages of superannuation etc., are offered, and applicants, who are asked to send us a curriculum vitae including present salary, in confidence, will be considered and offered remuneration according to their merit.

OXLEY DEVELOPMNETS COMPANY LIMITED, Priory Park, Ulverston, North Lancashire

MEDICAL RESEARCH COUNCIL Laboratory of Molecular Biology, Hills Road, Cambridge

which makes extensive use of on-line computer control requires ELECTRONICS ENGINEER holding HNC, or equivalent qualification, for design, construction and maintenance of electronic equipment.

The applicant should have experience of computer technology or digital electronics, or related fields.

Apply to the Administrative Secretary.

TELEVISION ENGINEERS

for outside servicing with experience in closed circuit medical, scientific, or allied applications required. A knowledge of 1" Helical scan V.T.R.s and colour television would be an added advantage. Company car provided. Salary according to experience. Any further information and interview.

SIEREX LIMITED, 15/18 Clipstone Street, London W1P 8AE. Telephone: 580 2464.

THE UNIVERSITY OF SUSSEX SCHOOL OF BIOLOGICAL SCIENCES

TECHNICIAN OR JUNIOR TECHNICIAN

Applications are invited for the post of Technician to build and maintain systems for the automatic control and recording of animal behaviour. Work will involve integration of electromagnetic and computer systems with simple mechanical and hydraulic equipment. Preference will be given to applicants with interests in either elementary programming, analysis or behavioural records, or running the experiments.

Salaries are in accordance with age, qualifications and experience within the ranges:

£373 to £575 for Junior Technicians £692 to £1007 for Technicians

In addition, qualification allowances are paid for approved qualifications. Three weeks paid holiday plus University Closures. Applications should be sent in writing to: Laboratory Superintendent, School of Biological Sciences, The University of Sussex, Falmer, Brighton, BNI 9QH, quoting ref. no. 135/2.

Commissioning Engineers and Installers TRANSMISSION DIVISION

In spite of an excellent response to our recent advertisement a continuing expansion projecting into the foreseeable future demands that we seek additional staff.

Opportunities exist for commissioning and installation staff with experience of carrier systems to join a well established team working on transmission contracts both in the U.K. and overseas.

The Transmission Division's growth and its heavy commitments also create openings for less experienced engineers with a good transmission background who would be prepared to accept responsibility after a period of field training.

If you meet any of the above mentioned requirements we shall be pleased to hear from you. Please telephone V. S. Klein, Installation Manager or write stating age and giving details of qualifications and experience quoting reference number BEE/393/E to the Personnel Officer, Personnel Department, The Plessey Company Limited, Beeston, Nottingham NG9 ILA tel. Nottm. 254831 Ext. 4497.

PLESSEY ELECTRONICS

APPOINTMENTS

ELECTRONIC TECHNICIANS

Marconi

Can offer you

NON-TIED HOUSING IN A NEW TOWN ATTRACTIVE SALARY **ANNUAL SALARY REVIEWS GOOD WORKING CONDITIONS 37-HOUR WORKING WEEK**

At Basildon we have a number of vacancies for technical test staff to work on advanced aeronautical electronic systems, maintenance and building of test equipment and other major projects. These positions will be of particular interest to men with experience of transmitters; receivers, aerials, closed circuit T.V. or digital systems.

Please telephone or write for an application form to :-

Mrs. B. Bridgen, Personnel Officer, The Personnel Dept., The Marconi Company Limited, Christopher Martin Road, Basildon, Essex. Phone: Basildon 22822.

It's Racal quality year'

And we are looking for good quality Service Engineers

to help us maintain our standards of Test Equipment service.

Specification:-

Wide general experience Good knowledge of circuit applications Experience with H.F. S.S.B. Communications Test Equipment.

Optional Extras:-City & Guilds or O.N.C. or H.N.C.

Power Consumption:-£1100 - £1300

Applications in writing please to:-

Mr. P. Cousins, Group Personnel Manager, Racal Electronics Limited, THE RACAL GROUP Western Rd., Bracknell, Berks.

Senior TEST ENGINEER-**TELEVISION RECEIVER** MANUFACTURE

Due to a promotion, Rediffusion Vision Service Limited, which is a large scale producer of Television Receivers, now require an experienced Test Engineer. He will be responsible to the Production Manager for taking charge of all fault-finding staff engaged in manufacture. Duties will include training and instruction of test staff and liaison with the Group Engineering Department.

EXPERIENCE AND QUALIFICATIONS

Full familiarisation with monochrome television techniques including the use of transistors is essential. Knowledge of colour techniques and experience in a similar position is desirable. H.N.C. or equivalent preferred but lack of formal qualification will not debar a suitable applicant.

SALARY

Subject to negotiation with a minimum of £1,500 per annum. Housing assistance and removal expenses are available.

Applications in confidence to:

Mr. J. W. Lunken, Manager, **Rediffusion Vision Service Limited,** Trading Estate, St. Helens Auckland, **Bishop Auckland, Co. Durham.**

A Member Company of the Rediffusion Organisation

SERVICE TECHNICIANS

Experienced Electronic Engineers, minimum qualifications O.N.C./City and Guilds or 2/3 years' Bench experience, to service and repair a wide range of electro-acoustic instruments. Driving experience essential. Excellent salary and opportunities for advancement.

Write or telephone for immediate interview:

APPOINTMENTS

Personnel Department, Amplivox Limited. Beresford Avenue, Wembley, Middlesex. Telephone 902-8991.

BROADCAST RELAY ENGINEERS

required for the

ISLAND OF MASIRAH

(Off the Coast of MUSCAT and OMAN)

For an unaccompanied tour of duty of I year preceded by about a month in U.K. for familiarisation, documentation and medical clearance.

Total emoluments in the range £2,436-£3,079

for service on the Island. Actual level within range will depend on experience and marital status.

Engineers experienced in the operation and maintenance of high-power broadcast transmitters and who are of Third Year City and Guilds Tele-communications Technical standard are invited to apply for full particulars to:

The Personnel Officer, **Diplomatic Wireless Service**, Hanslope Park, Wolverton, Bucks.

COMMUNICATION & CALL SYSTEMS **Speech & Visual**

Our steadily increasing volume of business, at home and overseas, now creates a requirement for additional engineering staff. We have immediate vacancies for Senior and Junior engineers with good practical experience in any of the following aspects of the work:

> System design. **Planning and Estimating.** Installation control. Test and commissioning.

The work is varied and interesting, with frequent opportunities for travel, and for contacts with other organisations. Applications, which will be treated in strict confidence, should be sent to:

The General Manager, Special Services Division, British Relay House, 41 Streatham High Road, S.W.16.

UNIVERSITY OF NOTTINGHAM THE LANGUAGE CENTRE PROGRAMME ASSISTANT

(male or female)

The Programme Assistant will be respon-The Programme Assistant will be respon-sible for providing multiple copies of master tape recordings for use in the language centre. The ability to organise his or her own work to meet the pro-gramme time table will be expected. Some typing ability is necessary, and an interest in tape recording would be an advantage.

Salary will be on the Technicians Scale of £692 to £1,057 per annum.

Applications in writing quoting the names of two referees to the Staff Appointments Officer, University of Nottingham, University Park, Nottingham.

TECHNICIAN

required, for work in laboratories and in connection with use of visual and aural aids, at CITY OF LONDON COLLEGE, Moorgate, E.C.2.

Interesting and varied duties.

Salary within range £750-£1,115 p.a. according to age (minimum 21) and qualifications. Local government pension scheme.

Further details and application form from the Secretary.

PORT OF LONDON **AUTHORITY Radio Technicians**

The Authority operate a complex telecommunications network which includes position fixing survey systems. V.H.F. and U.H.F. radio (both marine and shore-based). U.H.F. and micro-wave telemetric links, message switching and tape relay systems and low power real time digital computers. Staff are required to maintain this equipment at maximum efficiency and applications are invited from men interested in work which plays an important part in the smooth functioning of the Port.

Vacancies exist at Gravesend and King George V Dock, and successful candidates will be offered salaried positions on the Authority's permanent pensionable staff.

SALARY: £910 to £1.210 per annum. There are opportunities for promotion to a senior grade with salary up to £1.355 per annum. Commencing salaries will be in accordance with qualifications and experience. To ensure adequate coverage, a two-shift system is operated, for which an additional allowance is payahle.

★ Radar and Microwave Links
 ★ Digital and Telemetry

* Diskital Biol Telemetry Possession of ONC Electrical Engineering, City and Guilds Intermediate Certificate in Telecommunica-tions plus Radio II, or an equivalent standard of technical training in civil or service fields is desirable but not essential, and the ability to drive would be an asset. Practical training on specialist equipment will be provided where necessary be provided where necessary. Application forms may be obtained from:

The Chief Engineer (Personnel), Port of London Authority, P.O. Box 242, Trinity Square, London, E.C.3.

Additional authors are required for important projects at our London and Portsmouth offices, including on-site-opportunities, in the following fields:

 Data processing
 Solid state radar

 Servo systems
 Telecommunications

 Navigational aids
 Electronic instrumentation

 Sonar systems
 Electro-mechanical systems

Generous salaries are being offered, according to qualifications and experience. Formal qualifications to H.N.C. or equivalent, and a

Formal qualifications to H.N.C. or equivalent, and a minimum of three years in the engineering industry, will be an advantage.

Please apply in writing to:

The Technical Publications Manager, Irwin Technical Limited, 109-123 Clifton Street, London, E.C.2.

SALES ENGINEER CO-AXIAL CONNECTORS

Precon is growing rapidly and we are making new appointments to handle the already wide range of co-axial connectors for the electronics and communications industries. New models are being introduced—we have a vigorous expansion policy and the prospects for advancement are excellent.

Fringe benefits include car, pension scheme, and 3 weeks holiday.

Applicants should have experience in selling electronic components but be ready to become involved in wider commercial duties.

Apply in confidence to: Director and General Manager Precision Connectors Ltd.

56-58 Green Street, Forest Gate, London. E.7. Telephone: 01-552 3405. TRANSMITTER ENGINEERS £1,485-£2,365

We are looking for keen Engineers to join the Transmitter Section of our Station Design and Construction Department to assist with the heavy programme of work already under way to establish an extensive UHF network suitable for colour television. Most of our projects consist of three main stages:

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PLANNING—this involves consideration of specification requirements, tender appraisals, discussions with manufacturers and production of suitable layout drawings for transmitters and ancillary equipment.

CONSTRUCTION—during this stage it is necessary to hold regular meetings to ensure that the work is progressing in accordance with the planned programme, and to agree many detailed points not covered in the specification.

Finally there is the COMMISSIONING stage, when comprehensive works and on site acceptance tests are carried out to ensure the specification has been complied with before the station is handed over to our operations and maintenance engineers.

The work will cover all the above aspects. The appointment is based at our Knightsbridge, London, Headquarters but a considerable amount of travelling throughout the United Kingdom will be necessary for which appropriate allowances will be paid.

The successful applicants will need to have had some relevant experience with RF circuitry and television techniques. They should also have the ability to write clear and concise reports and to work on their own initiative. An HNC or equivalent qualification would be an advantage. Salary in above grade depending on qualifications, experience, etc. Excellent conditions of service.

Application forms from :---

The Personnel Officer INDEPENDENT TELEVISION AUTHORITY 70 Brompton Road, LONDON, S.W.3

APPOINTMENTS

Quote reference no. WW/1136/H.69/70.

Closing date for completed application forms: 7th May, 1969

LABORATORY TECHNICIANS

required by BBC in the Test Laboratory of Equipment Department, Chiswick, W.4.

The duties include the inspection, alignment and performance checking of equipment used in the Television and Radio Services. Applicants should have experience in a development laboratory, or be familiar with the problems encountered when new designs of electronic equipment, made by small batch production methods, are being tested. Preference will be given to applicants who have an H.N.C. or equivalent qualification in electrical engineering, but for those who have made some progress toward one, day release facilities will be given to complete their studies to this level.

Salaries on appointment will be between £1,215 p.a. and £1,550 p.a. in grades having maximum salaries of £1,560 p.a. and £1,775 p.a. Technicians lacking qualifications or sufficient experience may be appointed at a lower grade.

Request for application form to Engineering Recruitment Officer, Broadcasting House, London W.1A 1AA, quoting reference 69.E. 2081. W.W.

RANK PRECISION INDUSTRIES

test equipment calibration engineer

Skilled in the comprehensive calibration and minor fault correction of the usual range of commercial test equipment. Salary from £1,250—£1,400 per annum according to experience, etc.

The Rank Organisation

Holders of The Queen's Award to Industry for 3 successive years. of the Queen's Award to Industry for 3 successive years.

Apply to:

S RANK

INDUSTRIES

Personnel Manager Rank Precision Industries Ltd. Great West Road, Brentford, Middlesex. Tel: 01-560 1212

An immediate vacancy occurs at THE WIRELESS COLLEGE COLWYN BAY, NORTH WALES

for an additional instructor to assist in preparing students for P.M.G. examinations. The primary responsibility will be the theoretical instruction on the technical electricity section of the syllabus. Applicants must hold a P.M.G. Certificate and should have a sound theoretical knowledge. Recent Marine operating and/or teaching experience is desirable, but not essential. Write in the first instance to The Principal.

TECHNICAL AUTHORS

A Technical Publications Contractor has vacancies in their Home Counties offices and on site for personnel to be engaged in the preparation of manuals for a wide range of electronic and allied equipments. Applications are invited from practising or aspiring authors with relevant experience. Box No. 5056.

BATH UNIVERSITY OF TECHNOLOGY School of Chemistry and Chemical Engineering

EXPERIMENTAL OFFICER IN ELECTRONICS

Applications are invited for the above post. Duties include the design and construction of special purpose electronic equipment for research projects and the maintenance of electronic equipment within the School. The School has an on-line PDP8/K70 computer and it is intended to use this in conjunction with undergraduate teaching and post-graduate research projects. Although can-didates should have an interest in computer systems, previous experience with on-line computers is not essential, since training in relation to PDP8 computer maintenance can be arranged. The School also has several other facilities including Nuclear Magnetic Resonance, Mass Spectrometry, Infra-red and Utra-violet Spectrometry, Electron Spin Resonance Equipment and Chromatography. Applications are invited for the above post.

Experience in solid state electronics and modern construction and wiring techniques is essential.

Starting salary will be within the range £1,435-£1,715 per annum, for suitably qualified candidates.

Application forms from Registrar (S), The University, Claverton Down, Bath, quoting reference 69/20.

RADIO TECHNICIANS

Vacancies to be filled by October, 1969

A number of suitably qualified candidates are required for unestablished posts, leading to permanent and pensionable employment (in Cheltenham and other parts of the UK, including London). There are also oppor-tunities for service abroad. Applicants must be 19 or over and be familiar with the use of Test Gear, and have had practical Radio/Electronic workshop experience. Preference will be given to such candidates who can also offer "O" Level GCE passes in English Language, Maths and/or Physics, or hold the City and Guilds Tele-communications Technician Intermediate Certificate or equivalent technical qualifica-tions. A knowledge of electro-mechanical equipment will be an advantage. Pay according to age, e.g. at 19-£869; at

Pay according to age, e.g. at 19-2869; at 5-21,130. 25

Prospects of promotion to grades in salary range £1,217-£2,038. There are a few posts

carrying higher salaries. Annual Leave allowance of 3 weeks 3 days rising to 4 weeks 2 days. Normal Civil Service sick leave regulations apply.

Application forms available from:

Recruitment Officer (RT 3),

Government Communications Headquarters, Oakley, Priors Road, CHELTENHAM, Glos, GL52 SAJ.

ELECTRONIC ENGINEERS

Service Engineers required for Offices, throughout the United Kingdom, of well-known Company manufacturing Electronic Desk Calculating Machines. Applicants should possess a sound knowledge of basic Electronics with experience in Electronics, Radar, Radio and T.V. or similar field. Position is permanent and pensionable. Comprehensive training on full pay will be given to successful applicants. Please send full details of experience to the Service Manager, Sumlock Comptometer Ltd., 102/108 Clerkenwell Road, London, E.C.1.

COLOUR TELEVISION FAULTFINDERS & TESTERS

We have a number of vacancies in our Production Test Departments for experienced faultfinders and testers.

Knowledge of transistor circuitry and experience with Colour Receivers together with R.T.E.B. Final Certificate or equivalent qualifications required.

These will be staff appointments with all the expected benefits. Applications to:

Works Manager, Rediffusion Vision Service Ltd., Fullers Way South, Chessington, Surrey (near Ace of Spades). Phone: 01-397 5411

Pye Telecommunications Ltd. OF CAMBRIDGE

The largest exporters of VHF/UHF radiotelephone equipment in the world require

ENGINEERS AND DESIGN DRAUGHTSMEN

Type of work and experience: We require electronic engineers and design draughtsmen to join teams engaged in the design and development of fixed mobile and portable UHF and VHF transmitters and receivers. These teams are responsible for all aspects of designing and development through to the production line.

Applicants should have experience in economic design for quantity production in the same or similar field of activity.

Education. Appropriate degree or diplomas preferred or proven experience of comparable level will be considered.

Age: 20-40 years.

Company contribution Pension Scheme.

Applications should be submitted to PERSONNEL MANAGER

CENTRAL MIDDLESEX GROUP HOSPITAL MANAGEMENT COMMITTEE

NEW POST-ULTRA-SONIC TECHNICIAN

This is the first post of its kind and will attract a man with a strongly developed interest in electronics who wishes to consider the application of ultra-sonics to medical examinations.

The successful candidate will be required to work on his own initiative, be capable of handling patients and prove able to conduct ultra-sonic examinations single-handed on occasions. He will also be required to develop, under supervision, new electronic apparatus.

Possession of a car is essential and a car user allowance will be payable.

Further details and application forms available from: Group Secretary, Central Middlesex Hospital, Park Royal, N.W.10.

ELECTRONICS DESIGN ENGINEERS SENIOR DESIGN DRAUGHTSMEN

are required to work on a variety of challenging problems in a rapidly expanding company. An ability to assume a large degree of individual responsibility as part of integral design team is required or will be encouraged and developed.

Our design projects include micro-electronics, digital computers, static inverters, power supplies and complete Systems Designs both Ministry and Commercial. Realistic salaries will be proportional to general ability. For further details and interview for either position please reply in writing to:---

Personnel Officer, GRESHAM LION GROUP LIMITED, TWICKENHAM ROAD.

HANWORTH, MIDDLESEX.

the theoretical principles of and experience in the maintenance of at least FOUR of the following groups of Communications, CMA Navigational and Surveillance Systems.

 Medium powered H.F. Transmitters and associated Receivers: Frequency Shift Keying, S.S.B. and D.S.B.

Bernes Comme Comme Comme Comme Comme

Apply to CROWN AGENTS, M. Dept., 4, Millbank, London, S.W.I., for application form and further particulars, stating name, age, brief details of qualifications and experience and quoting reference M2Z/690315/ WF

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UNITED SHEFFIELD HOSPITALS ROYAL INFIRMARY, SHEFFIELD S6 3DA

ELECTRONICS TECHNICIAN

Applications are invited for the position of Electronics Technician, to work on a research project in the University Department of Surgery in association with the Regional Medical Physics Department. The Technician will work in an electronics laboratory on design and development of equipment to be used in the clinical research. Applicants are expected to have O.N.C. or H.N.C. in Applied Physics or Electrical Engineering or equivalent qualifica-tions, with some practical experience in electronics. Salary will be on the Medical Physics Technician Scale V or IV (£711 to £1,050) depending on

qualifications and experience.

Applications giving names and addresses of two referees, to the Superintendent quoting Ref. 137.

Independent Specialist Radio, Television, Audio, Tape, etc., dealer

requires a young engineer, preferably minimum C. & G. Inter., to reinforce staff of increasingly busy Service Department. A conscientious and adaptable extrovert able to make good use of a comprehensive range of test gear and having a clean driving licence will be rewarded on a five-day week basis and will have the pleasure of working with similarly orientated colleagues.

Write in confidence to Managing Director, The Studio (Abingdon) Ltd., The Square, Abingdon, Berkshire, glving full background details and indicating availability for interview. All applications will be answered. [169

Technical Author A vacancy within the Publicity Department of the Radio Systems Division has arisen for a Technical Author. Applicants, whose minimum qualifications should include O.N.C. or equivalent, will ideally have a knowledge of radio and electronic engineering. The successful applicant will be required to assist in the writing of technical sales proposals on the systems and equipment manufactured by the Division. Assistance in this task will be given by a creative team of engineers, artists and printers.

Generous salaries according to experience and qualifications will be negotiated and there are excellent staff benefits within this progressive Company.

Applications, giving a brief career history, and quoting Ref. ILF | 755 | E to: The Technical Staff Manager, The Plessey Company Limited, Ilford, Essex

PLESSEY ELECTRONICS

ELECTRONICS DESIGN DEVELOPMENT ENGINEERS

Needed for work on advanced systems for voltage and current regulation at large power levels. Applicants should have H.N.C. as minimum qualification, and preferably some industrial experience. Assistance with housing is available if needed.

LABORATORY TECHNICIAN

Needed to assist in the development of voltage regulating and allied equipment. The successful applicant will have at least O.N.C. in Power Electrical Engineering (or an equivalent qualification), and will find some familiarity with simple electronics and closed loop control principles useful. As a member of a small but enthusiastic team, working in a rapidly changing technology, he will need to show drive and initiative which he will find well rewarded both by job interest and by a salary properly matched to his qualifications and abilities. Housing assistance is available.

Write with career details to:---

The Technical Director, **BRENTFORD ELECTRIC LIMITED.** Manor Royal, Crawley, Sussex.

PRODUCTION TEST ENGINEERING

Due to our successful Research and Design work many exciting new projects are entering a production phase and we require Engineers and Technicians to participate in this work.

Minimum qualifications required are a basic understanding of Transistor circuitry enabling testing to specification to be carried out on our Data Processing and Servo Control Systems, etc.

Electrical Engineering Certificates an advantage, but not essential if experience in a similar activity can be offered.

Apply:

Personnel Officer **RECORDING DESIGNS LTD. Blackwater Station Estate** Blackwater, Camberley, Surrey **Telephone Camberley 24622**

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SEISMIC OBSERVERS

with analogue or digital field experience required for overseas service on land or sea, by

GEOPHYSICAL SERVICE INTERNATIONAL

who offer a good salary and foreign bonus, ample leave on full pay and foreign bonus, medical insurance scheme, life insurance, profit sharing and a pension plan. Those interested please write to:

The Personnel Manager Geophysical Service International Ltd. Canterbury House, Sydenham Rd., Croydon, Surrey quoting ref. 12/68, or telephone 01-686 6511

SPACE RESEARCH

THE EUROPEAN

requires for its Sounding Rocket Launching Range (ESRange) at Kiruna (Sweden)

Head of Operations Division

The holder of this post is expected to be a qualified engineer with univ. degree or equivalent (preferably in electronics or aeronautics) and with 5 to 10 years' experience in sounding rocket operations or missile testing.

His activities will cover: launching forecasts, technical information and liaison with users of the Range, preparation, co-ordination, checking, and reporting of launching procedures, maintenance and preparation of launching facilities and rockets, and directing recovery.

He will be in charge of app. 10 engineers and technicians and directly responsible to the Head of ESRange,

For the Instrumentation Division

2 electronic engineers with univ. degree or equivalent and appropriate prof. experience.

1. Head of Division

Responsible to the Head of ESRange for utilisation, maintenance and planning of a £2m. installation and in charge of a staff or 40 electronics engineers and technicians. He must have relevant experience in management and of electronic equipment systems including:

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