# Wireless World 

## BLECTRONICS Radio . Television



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## weight

＊It withstood a $2 \frac{1}{2}$ cwt．shock drop of 20 ft ． out of water；towed the camera at speeds up to 12 knots；raised and lowered the camera at 250ft．per minute over narrow diameter pulleys and capstan．

Subsequent examina－ tion proved the cable and coupling to be completely waterproof，resistant to twisting and electrically intact．

## CABLES






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## Wirelless World

ELECTRONICS, RADIO. TELEVISION

Managing Editor: HUGH S. POCOCK, m.I.E.E.<br>Editor:<br>H. F. SMITH<br>Assistant Editor: F. L. DEVEREUX, b.sc.

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A high gain transistor preamplifier is described which uses a single OC70 in grounded emitter connection. The supply voltage is 250 V and the load more than $330 \mathrm{k} \Omega$. The basic circuit arrangement is shown at ' A '.
The preamplifier has distinct advantages over a thermionic valve preamplifier. First, two valves would be required to get comparable gain. Second, thermal noise is sufficiently low for it to approach that of a valve circuit, and the transistor preamplifier is of course completely free from hum and microphony. The current drain is low, about 0.7 mA from the 250 V supply.

This preamplifier has comparatively high output impedance and best results are obtained when it is fed into the high input impedance of a thermionic valve amplifier. It can be used with low impedance microphones, pickups and tape heads. For a nominal circuit, with an input voltage of 5.5 mV and a gain of about 330 , the output voltage is about 1.8 V . The input impedance at $\mathrm{X}-\mathrm{X}$ is $200 \Omega$ and the output impedance at $\mathrm{Y}-\mathrm{Y}$ is of the order of $5 \mathrm{k} \Omega$. The frequency response depends to a certain extent on the source impedance. With a $50 \Omega$ source a 3 dB reduction in gain is reached at $15 \mathrm{c} / \mathrm{s}$ and $12 \mathrm{kc} / \mathrm{s}$.

Although designed for a 250 V supply voltage the preamplifier operates usefully with supply voltages down to as little as 100 V . It can be used successfully with a $30 \Omega / 50 \Omega$ microphone and a 100 V supply coming from a valve amplifier into which it feeds.

The preamplifier can be fed from the same supply as a valve amplifier by rearranging the circuit as at ' $B$ '. In the setup shown at ' $B$ ' one side of the output is earthed, whereas in ' $A$ ' the input and output are 'floating' at some voltage above the chassis. The input is fed between base and emitter through R1 and C1 so that R5 does not contribute a.c. negative feedback but forms part of the load. The output in ' $A$ ' can be taken from between C2 and chassis if desired, but it then becomes slightly smaller than when taken from C2-C3, because R5 no longer forms part of the load. Although the input terminals in both ' A ' and ' B ' are 'floating', there should be no risk of hum being introduced provided the preamplifier is mounted reasonably close to the microphone or pickup.

If the circuit is used in arrangement ' $A$ ' care must be taken not to short the input terminals to earth. In the arrangement shown at ' B ' the current through the pickup or microphone will not be excessive if the input is accidentally shorted to earth.

High voltage gain is obtained from a transistor in the same way as for a pentode valve, by operating it with a high load and feeding from a high supply voltage. In both ' $A$ ' and ' $B$ ' sufficient d.c. stabilisation is provided by the potential divider R2-R3 and emitter resistor R5 to ensure satisfactory operation up to an ambient temperature of $45^{\circ} \mathrm{C}\left(110^{\circ} \mathrm{F}\right)$. Besides the d.c. feedback provided by R2, R3 and R5, there is an a.c. feedback path formed by R2, R1 and the source impedance, all
in series. Part of the a.c. voltage developed across this potential divider is tapped off and fed back into the base. By including a $100 \Omega$ resistor R1 in the a.c. feedback path, effective feedback is ensured even when the source impedance is very small. Apart from improving the frequency response and reducing distortion, the a.c. negative feedback decreases the input and output impedances. If thought desirable a.c. feedback can be prevented by making R2 up of two nearly equal resistances and bypassing the common point to ground by a suitable capacitance, of the order of $0.5 \mu \mathrm{~F}$.

Even with a supply voltage of 250 V the circuit is so designed that the collector to emitter voltage never exceeds the d.c. voltage rating of $\mathrm{V}_{\mathrm{c}} \mathrm{max}=-5 \mathrm{~V}$ for the OC70 under the worst possible combination of conditions. The effect of resistance tolerances, supply tolerances, change of ambient temperature, and spread in transistor characteristics have all been considered in the design. In order to keep available the maximum voltage across the transistor, while allowing a tolerance of $\pm 10 \%$ on the supply voltage, the resistor tolerances must be within $\pm 5 \%$ to prevent the 5 V rating from being exceeded. High stability resistors should be used. For a nominal circuit the collector current is 0.7 mA and the collector to emitter voltage approximately 4 V . An OC70 with a low current amplification factor $a^{\prime}$ and low leakage current $\mathbf{I}_{\text {co }}^{\prime}$ (o) in grounded emitter causes the highest collector to emitter


The circuit is similar to that deseribed bv
The circuit is similar to that described bv
James J. Davidson in an article called James J. Davidson in an ar ficle called
'High Gain Transistor Amplifier, pub'High Gain Transistor Amplifier,' pub-
lished in Audio, October 1955 . voltage of 5 V . A transistor at the other extreme gives a collector to emitter voltage of about 0.8 V at the maximum ambient temperature of $45^{\circ} \mathrm{C}$, and allows an output of about 400 mV (r.m.s.).

If resistors of $\pm 10 \%$ tolerance are used, the collector to emitter voltage should be metered to ensure that it does not exceed 5 V . The performance is the same as for resistor tolerances of $\pm 5 \%$ except that possibly less collector voltage swing will be available if all the resistors have their lowest extreme value.

# Wireless World 

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## 625 Lines Again

IN the August issue of Wireless World we printed an article comparing, in the simplest possible terms, the picture quality offered by the various television systems that have been adopted in different parts of the world. The 625 -line came out rather badly from that comparison, the conclusion being that better value could be obtained for the bandwidth employed. Before that article was published there was a body of opinion strongly in favour of adopting the continental standards for the projected British colour service. Since then, enthusiasm for 625 lines has waned notably.

However, there are still those who consider the benefits of standardization outweigh other considerations. Among these is Sir Robert Renwick, who again writes in our correspondence columns this month to urge the advantages of 625 lines in the interests of the export trade.

Is it possible to reach a compromise that will satisfy the requirements of standardization and at the same time overcome the technical shortcomings of the normal continental system? Such a compromise might quite easily be reached by modifying the standard 625 -line system by widening the video bandwidth by about $1 \mathrm{Mc} / \mathrm{s}$. That would provide for a picture of sensibly equal horizontal and vertical definition and the slight deviation from complete standardization should be no great embarrassment to the export trade. If we are in fact to have colour television incompatible with the present service, can we do better than tentatively adopt these modified 625-line standards?

We say "tentatively" because these discussions of definition standards for colour television seem to become more and more unrealistic as one considers the other and far more pressing problems that have to be tackled before practical receivers can be produced at an economic price. These problems were frankly discussed at the official opening of the Sylvania-Thorn colour television laboratories at Enfield on October 3rd. The spokesman of both the American and British companies co-operating in this new enterprise expressed open dissatisfaction with colour techniques as they exist today, and the opinion was expressed that an entirely new system was needed.

A new series of colour test transmissions by both the B.B.C and I.T.A. was announced recently, and it was stated that these would be on both 405 and

625 lines. This seems to offer an opportunity for trying the wide-band 625-line modification to which we have referred.

## Power by Radio

THE centenary of the birth of Nikola Tesla is being celebrated this year in his native Yugoslavia and in the United States, his adopted country. Tesla made many important contributions to electrical engineering; in radio, he is mainly remembered for his attempts to transmit power by wireless. Looking back, it is safe to say he would have achieved greater success by concentrating his efforts on the substance of radio communication instead of on the shadow of power transmission. After more than half a century, that objective is still almost as far from attainment as in Tesla's day.
But the modest power requirements of the transistor have made us change our ideas as to what is worth while in the way of power transmission. Very opportunely coinciding with the Tesla centenary comes an article in an American journal* on the use of radio-transmitted power for the operation of transistorized receivers, navigational aids and other electronic equipment. Though important practical applications of the idea do not readily come to mind, it is of some academic interest.

For its simplest application "free power", as Dr. Hollmann calls it, is obtained by rectifying the carrier of a nearby broadcasting station with a crystal diode. The rectified and filtered output is then fed to transistors, which amplify and detect the signals of more distant stations.

A still more intriguing possibility is to make the wanted station supply the power for amplifying its own signals. At first sight that sounds impossible, but Dr. Hollmann claims that a receiver working in that way does in fact give a performance superior to that of a plain non-amplifying crystal set. The explanation he gives is that the plain crystal receiver derives its output power from the modulation envelope only; the carrier energy is wasted. In the amplifying receiver, on the other hand, the carrier is rectified to provide power for the amplifier, and so a greater output is obtainable.

[^0]
# Personal Paging System 

SELECTIVE INDUCTION METHOD

THE advantages offered by transistors in the design of small, lightweight equipment of low power consumption are particularly well emphasized by the new selective paging system recently installed at St. Thomas's Hospital, London. Here each member of the staff on call is equipped with a small transistor "receiver" which clips into a top pocket like a fountain pen (Fig. 1). It measures $5 \frac{3}{8}$ in long by 1 in diameter and weighs only $50 z$. The current consumption from the $1.35-\mathrm{V}$ mercury cell is only 0.5 mA in the quiescent condition (the receiver being permanently switched on) and 3 mA when a call is being received, and this is reckoned to give a cell life of over two months.

Developed by Multitone in collaboration with St. Thomas's, the paging system works on the magnetic induction principle. A single loop of wire round the building carries audio frequency signals from a 70 -watt amplifier and the magnetic field so generated is picked up in each "receiver" by a small coil with a laminated core (see Fig. 2). The system is made selective by assigning to each "receiver" a particular audio frequency which is transmitted as a call signal. There are 56 of these frequencies available altogether, in the range of $2-15 \mathrm{kc} / \mathrm{s}$, and each "receiver" selects its own frequency by means of a high-Q tuned circuit. Reception of the call signal changes the bias of a local transistor a.f. oscillator, which emits an audible
$1,000-\mathrm{c} / \mathrm{s}$ note through a miniature telephone earpiece in the "cap" of the "fountain pen."

In the transmitting equipment (Fig. 3) each call frequency is selected by pressing a combination of buttons. This operation also causes the signal to be automatically coded, by a gating pulse generator, into a pattern of marks and spaces which means something to the recipient at the other end when he hears the resulting bursts of $1,000-\mathrm{c} / \mathrm{s}$ tone. The duration of the call signal is also determined automatically. Usually the code signal means "go to the nearest telephone to receive a message." The equipment is, however, capable of conveying speech signals in addition to the code signals Here, a code signal is transmitted first which means "prepare to receive speech." The recipient then presses a button on the side of the "pen," which puts the high-Q selector circuit out of action (and also the $1,000-\mathrm{c} / \mathrm{s}$ oscillator), and he places the instrument near to his ear.

At the transmitter a "speech" key is operated which takes out of circuit the pulse generator producing the code sequences and switches in a crystal ${ }^{-}$ microphone and amplifier. The tendency for the speech signals (which, of course, cover a wide band) to trigger the selector tuned circuits of other "receivers" is avoided by transmitting the speech


Fig. 3. The transmitting equipment and control panel.

on lower power. This occurs automatically because the loop is tuned for the individual code signal frequencies but not for speech.

The construction of the "fountain pen" receiver can be seen from Fig. 2. At the bottom end is the mercury cell, then the pick-up coil and the 3-stage transistor circuit which amplifies the
induced currents. Transformer coupling is used between stages and all components are miniature types except the trimmer capacitor for tuning the high-Q selector circuit. The coil for this tuned circuit is enclosed in a ferrite "pot" core, built into the "cap" of the "fountain pen," and has a Q of about 30 .

## "A Restrictive Practice?"

UNDER this heading in our January, 1954, issue we announced that the question of the supply of valves and cathode-ray tubes was to come under the scrutiny of the Monopolies and Restrictive Practices Commission. It was stated in the official communiqué that the Commission was to "report about both the facts of the matter, and their bearing on the public interest."

In August this year, however, it was announced that the Commission's findings "are now to be a factual report only." All such reports must, of course, be factual, but the significance of this phrase is that it will describe the industry's arrangements, but will not give any views on whether or not they are in the public interest.

The report was received by the Board of Trade on September 17th, and "will be published in due course."

On October 4th the British Radio Valve Manufacturers' Association (B.V.A.)* issued a statement, which we quote in full below, announcing changes in the constitution and trading practices of the Association having effect from September 1st.
"It will be known that the Association abandoned its 'stop list' and allied provisions some years ago and it has now discontinued all arrangements for collective resale price maintenance on the part of the manufacturers. This has been made necessary by the Restrictive

[^1]Trade Practices Act which, in effect, prohibits this practice. It will in future be for each individual manufacturer to maintain the prices of his own valves and tubes if he so desires.
"Hitherto 'B.V.A. Prices' have been fixed by agreement of all manufacturers who are members of the Association. This policy has also been abandoned. From knowledge of the structure of the industry, however, and in view of the present period of recession with continually rising costs, it will be appreciated that although prices are no longer to be fixed by agreement it does not follow that the prices of comparable valves and tubes will necessarily vary between one manufacturer and another in the immediate future.
"In coming to the decision to abandon collective fixing of prices, the Association has had in mind that if this practice were to be continued it would in all probability have to be justified in the very near future before the Restrictive Trade Practices Court in the light of the narrow economic criteria set out in the Act. The practice of fixing prices by agreement is not in the present state of the industry of the same degree of importance as it has been in the past or as it may well be in the future.
"Apart from the foregoing the Association is continuing its general policy in the interests of the public, the trade and the industry itself, including its vast field of technical collaboration between the manufacturers with the Services and in international fields."

## Continuity of Power Supply

## Bridging the Gap Between Mains Failure and the Connection of an Auxiliary Source

IN all essential services normally dependent on power from the public electricity supply, it is customary to have a standby source available for use in case of failure of the mains. This may be a bank of storage batteries or an auxiliary generating plant, but in either case there will be a break in the supply while automatic chañgeover switches are tripped or the engine is started and run up to speed.

To overcome these difficulties a motor-generator power plant developed by Standard Telephones and Cables in conjunction with Pelapone Engines, Led., of Derby, incorporates a heavy flywheel which stores enough kinetic energy not only to start a standby diesel engine but also to drive the loaded a.c. generator

[^2]until the engine has developed full power. A magnetically operated clutch between flywheel and engine is arranged to engage when the external power supply is interrupted.

# WORLID OF WIIRELESS 

Organizational, Personal

and Industrial Notes
and News


NORTHERN service area of the I.T.A. when Emley Moor joins Winter Hill in November.

## I.T.A. Progress

THE fourth I.T.A. transmitter (on Emley Moor, near Huddersfield, Yorks.) opens on November 3rd -within 14 months of the opening of the Authority's first station.
It will use a directional aerial, vertically polarized, with an e.r.p. varying from 10 to 200 kW - the flat side of the polar diagram being towards the Pennines to avoid undue overlapping of the service areas of the two northern transmitters. It will operate in Channel 10 ( $199.75 \mathrm{Mc} / \mathrm{s}$ vision, $196.25 \mathrm{Mc} / \mathrm{s}$ sound).
It is hoped that it will be possible to transmit trade tests on high power for about a week before programmes begin. The present low-power pilot transmissions will continue to be broadcast daily, except on Sundays, until then.

## V.H.F. Receiver Response

READERS in south-east England interested in knowing the frequency response of their v.h.f. sound receivers will be pleased to learn that the B.B.C. has introduced standard-frequency test transmissions from Wrotham. They are being radiated each Thursday afternoon from 4.45 until 5.10 by- the Third Programme transmitter ( $91.3 \mathrm{Mc} / \mathrm{s}$ ). The B.B.C. hopes to maintain the level of tone accuracy within about $\pm 1 \mathrm{~dB}$ after allowing for $50-\mu \mathrm{sec}$ preemphasis and de-emphasis. The depth of modulation, excluding pre-emphasis, is 30 per cent, with the exception of the frequency marked with an asterisk which is 100 per cent.

The schedule is:-1645-1646, plain carrier; 1646$1648 \mathrm{lkc} / \mathrm{s}^{\star}$; $1648-1650,1 \mathrm{kc} / \mathrm{s} ; 1650-1653$, halfminute transmissions of each of the following frequencies $50,100,440,1,000,5,000$ ànd $10,000 \mathrm{c} / \mathrm{s}$; 1653-1703, plain carrier; 1703-1710, as from 16461653.

## B.S.R.A. Show

THE 1957 convention and exhibition of the British Sound Recording Association will be held in the autumn (September 20th-22nd), instead of in the spring as in previous years, and will again be at the Waldorf Hotel, Aldwych, London, W.C.2.

The Exhibition, which will be similar in general character to the eight previous shows organized by
the Association, will emphasize the technical and engineering aspect of the equipment to be shown.
In addition to providing individual demonstration rooms for exhibitors, there will also be a communal demonstration hall.

## "Importing an Instrument"

SINCE the publication of A. J. Reynolds' article under this heading in the October issue the Board of Trade has issued a revised form of application for the Duty Free Licence, which makes that part of the article referring to these licences meaningless.
The new form is in effect designed to make life so difficult for the applicant that he gives up the struggle and pays the duty. Question 8, for example, asks for copies of all correspondence with U.K. manufacturers during the search for an alternative.
Question 9 asks for the precise grounds on which exemption from duty is claimed, and, if superiority of performance is claimed, demands supporting evidence (in duplicate!) which "may be referred to the British manufacturers for their observations."

The new official attitude goes so far as to refuse duty-free importation of spares for instruments already in the country (and imported duty free) if a "similar" British instrument is now available.

## PERSONALITIES

Sir Stanley Angwin, K.B.E., has retired from the chairmanship of the Commonwealth Telecommunications Board which he has held since 1951, and was succeeded on October 1st by Sir Ben Barnett, K.B.E., who recently retired from the Post Office where he was deputy director-general. Sir Stanley, who is 73, was Post Office engineer-in-chief from 1939 until his resignation in 1946 on his appointment as chairman of Cable and Wireless, which he relinquished on joining the Commonwealth Telecommunications Board. He was chairman of the Radio Research Board from 1947 to 1952 and a year later was awarded the Faraday Medal by the I.E.E.

Sir Lawrence Bragg, F.R.S., a director of the Davy Faraday Laboratory of the Royal Institution, will give the third Clerk Maxwell memorial lecture during the 195.7 convention on "Electronics in Automation" being organized by the Brit.I.R.E. The lecture will be given on the first day of the convention, June 27th, in the Cavendish Laboratory, Cambridge.

In recognition of "his contributions to the advancement of radio science, and in particular for his long series of notable editorial articles in Wireless Engineer," Professor G. W. O. Howe, D.Sc., LL.D., Wh.Sch., M.I.E.E., has been elected an honorary member of the British Institution of Radio Engineers. Having been professor of electrical engineering at Glasgow University for 25 years (1921-1946), it is fitting that the proposal for his election should come from the Scottish section of the institution. He becomes the 15th honorary member of the institution, and the fourth elected since the end of the war.

Dr. Balth. van der Pol is due to retire from the directorship of the International Radio Consultative Committee at the end of the yeat. At the recent C.C.I.R. plenary assembly in Warsaw, Ernest Metzler, head of the radio section of the Swiss P.T.T., was elected to succeed him. Dr. van der Pol was appointed director of the C.C.I.R. in 1948, and his term of office was extended by two years to cover the recent plenary assembly. In 1953 he was awarded the Valdemar Poulsen gold medal by the Danish Academy of Technical Sciences for his work on the propagation of radio waves both in theory and practice. Prior to joining the C.C.I.R. he was a member of the board of the Physics Laboratory of the Philips organization in Eindhoven, Holland.
H. E. M. Barlow, B.Sc.(Eng.), Ph.D., M.I.E.E., Pender Professor of Electrical Engineering at University College, London, has joined the editorial advisory board of our sister journal Wireless Engineer. Professor Barlow has been a member of the academic staff of the Faculty of Engineering, University College, since 1925. He joined the Telecommunications Research Establishment in 1939, and in 1943 was appointed superintendent of the radio department at R.A.E., Farnborough.
"For their work in originating and developing the Decca Navigator System" William J. O'Brien and Harvey F. Schwarz have jointly been awarded the Gold Medal of the Institute of Navigation. Mr. O'Brien conceived the system of hyperbolic phase-comparison navigation in the United States in 1938, and with the help of Mr. Schwarz, then of the Decca Record Company, arranged for demonstrations in this country which culminated in its operational use as a mine-sweeping aid in the D-Day landings. There are now Decca chains covering the whole of the British Isles and most of the Western European seaboard.
O. W. Humphreys, B.Sc., F.Inst.P., M.I.E.E., director of the G.E.C. Research Laboratories, Wembley, has been elected president of the Institute of Physics. He has been a vice-president since 1952 and has been chairman of the International Special Committee on Radio Interference (C.I.S.P.R.) for the past three years.
A. J. Brunker, B.Sc.(Eng.), A.C.G.I., D.I.C., A.M.I.E E., chief engineer of E. K. Cole, Limited, which he ioined in 1947, has been appointed an executive director of the company. Before joining Ekco he was at the Ministry of Supply where for several years he held the post of deputy director, radio production. Mr. Brunker is also a director and general manager of Ekco Electronics, Limited.
A. Berkovitch has been appointed manager of Philips' car radio department in place of A. F. D. Knight, who has now taken up another appointment within the company's television and radio division. Mr. Berkovitch, who was with Philips for a short time before the war, was for nine years in charge of telecommunications and transport with the International Police Force in Trieste.
J. M. Bedford, A.M.Brit.I.R.E., until recently chief engineer of the radio section of Igranic, Ltd., has joined the research and developing department of Parmeko, Ltd., manufacturers of transformers, transductors and magnetic amplifiers.
T. C. Isaac, who joined Ambassador Radio and Television, Limited, as chief engineer in 1953, has been appointed technical director of the company. Before joining Ambassador he was chief engineer of Mains Radio-Gramophones, of Bradford, the manufacturing company for Radio Rentals, Limited. K. H. Yandell, who joined Ambassador this year, has also been appointed to the board as sales director. He has previously been with Philco, Regentone and R.G.D.
C. G. Allen, who has been with McMichael Radio for over 33 years, has ceased to be a director and sales manager of the company. He hopes to retain a connection with the radio industry.

## OUR AUTHORS

H. de Laistre Banting, who writes on the design of multi-standard television receivers in this issue, has been in charge of the television department of the Société Belge de Télécommunications in Brussels since 1954 where he has been concerned with the development of receivers for the Benelux countries. Before going to Belgium he was for several years in the research department of Murphy Radio, having previously been with Bush Radio which he joined in 1941.
J. D. Smith, author of the article on recording characteristics, graduated in 1950 at Bristol University with second class honours in physics. After two years graduate apprenticeship with the General Electric Company at their radio and television works, he joined the staff of the company's applied electronics laboratories. His interest in sound reproduction is purely a leisure-time pursuit.
P. Tharma, who describes a transistor receiver in this issue, joined the transistor section of Mullard's valve measurement and applications laboratory last year. He was educated at Ceylon Technical College where he received a B.Sc. honours degree in engineering in the external examination of the University of London. On coming to this country he received his telecommunications training with the G.P.O.

## OBITUARY

As a result of a motoring accident, Roland Harris Dunn, B.Sc., A.M.I.E.E., head of an advanced development section of the Telephone Division of Standard Telephones and Cables, died on September 7th at the age of 49. After graduating at Manchester University he was for a short while with B.I. Callender's Cables before joining the engineering staff of S.T.C. in 1930. In recent years he has specialized in various branches of electronics with particular reference to telemetering.

Maurice C. Jones, A.M.I.E.E., general manager of Gardners Radio Limited, of Somerford, Christchurch, Hants, for the past ten years, died in August aged thirty-nine. Before joining Gardners he was in the Telecommunications Research Establishment at Langton Matravers and later at Malvern.

## IN BRIEF

As was anticipated, television licences reached the six million mark during August. The month's increase was 64,820 , bringing the total at the end of the year to $6,044,330$. The total number of sound receiving licences at the end of the year was $8,038,062$, including 307,082 for car radio. The overall total of broadcast receiving licences current in the United Kingdom at the end of August was, thercfore, $14,389,474$.

The date of the Radio Industry Council's proposed Scottish Radio Show has been announced. It will be held at the Kelvin Hall, Glasgow, from May 22nd to June 1st next year. Although this is the first show to be organized in Scotland by manufacturers since 1935, the Scottish Radio Retailers' Association has held two Glasgow shows since the war.


PRODUCTION LINE at Ferranti's new Manchester works where Pegosus (shown here) and Mercury computers are being manufactured in quantity.

## London Computer Group

 has been formed to enable people with different professional interests to examine, through study groups, common problems in computer use. Membership is open to anyone with an interest in computers. The joint honorary secretaries are at 19A, Coleman Street, London, E.C. 2 (Tel.: Monarch 7822).An electronic instruments exhibition is being staged by E.M.I. Electronics, Ltd., at the Royal Hotel, Woburn Place, London, W.C.1, from November 28th to 30th. In addition to some one hundred instruments being shown and demonstrated, it is also proposed to show computers for machine tool control in operation. The exhibition will be open from 10 a.m. to 6 p.m., and admission is by ticket obtainable free on written application to E.M.I. Electronics, Ltd., Hayes, Middlesex.

Membership of the Brit. I.R.E. increased during the year ended March 31st by 310, bringing the total to 5,392.

The R.I.C. panel of judges, on which Arthur Clarkson (G.E.C.) recently replaced W. M. York (E. K. Cole), is considering articles already submitted for the 1956 premiums for technical writing. Articles should be submitted to the R.I.C., 59, Russell Square, London, W.C.1, before December 31st.

A short course of evening lectures on the microwave behaviour of ferrites begins on October 16th, at Sir John Cass College, Jewry Street, Aldgate, London, E.C.3. The fee for those residing in the administrative county of London is $£ 1$.

Two courses on transistors have been organized by the Borough Polytechnic, Borough Road, London, S.E.1. The first, covering basic principles, began on October 9 th, and the second, on special applications, begins on January 15 th. The lectures will be given in the afternoon and repeated in the evening. The fee for each ten-lecture course is 25 s .

Demonstrations of "hi-fi" equipment and stereophonic tapes are being given by Classic Electrical Co., L.td., at the Croydon Civic Hall on November 26th and 27th. On the first day demonstrations will be given from 1.0-2.0, 6.30-8.0 and 8.30-10.30, and on the second day in the evening only.
S.I.M.A. Officers.-At this year's annual general meeting of the Scientific Instrument Manufacturers' Association, G. A. Whipple, M.A., M.I.E.E., F.Inst.P., chairman and managing director of Hilger and Watts, was installed as president. P. Goudime, of Electronic Instruments, was elected vice-president and the following as new members of the council: F. W. Dawe (Dawe Instruments), P. J. Ellis, O.B.E. (Pullin), J. M. Furnival, M.B.E. (Marconi Instruments), D. F. Newstead (Rank Precision Industries), J. A. Stafford (Taylor, Taylor and Hobson), W. H. Storey (Unicam Instruments) and N. Trepte (Griffin and George).
T.E.M.A.-The annual report of the Telecommunication Engineering and Manufacturing Association, of which H. Faulkner is director, deals largely with Post Office matters of policy and development as they affect the telecommunications manufacturing industry. In the section dealing with education and training in the industry it is announced that the Association is preparing a careers handbook for publication later this year.

Amateur Exam. Results.-Of the 518 candidates who sat for the radio amateurs' examination conducted by the City and Guilds of London Institute in May, 458 ( 88.4 per cent) were successful. Four of the candidates were blind and two bedridden; for these special arrangements were made for the tests to be taken at home.

Instruments-measurement and control-and com'ponents associated with them are classified under some 2,000 headings in the buyers' guide section of the 1956 edition of "The Instrument Directory" issued by Instrument Practice. The 244-page directory includes lists of manufacturers, trade names and industrial associations.

The annual report of the Institute of Physics records that during the past year the membership increased to over 5,000 . The Institute's recently established graduateship examination was taken in four centres by 54 candidates, of whom only 11 passed.

## BUSINESS NOTES

The Mervac Printer, an exposing unit produced by Grant Production Company (4, Rathbone Place, Oxford Street, London, W.1), can be used in a dual rôle in the production of printed circuits. It can produce the film negative from the original wiring diagram, and also print the circuit on the copper laminate. The company also provides a service for the production of printed circuits to manufacturers' requirements.

Webcor (Great Britain), Ltd., formed last year as a subsidiary of the American Webster-Chicago Corporation to market in this country record players and receivers, has temporarily suspended operations. Enquiries should be sent to Belcher (Radio Services), Ltd., 59, Windsor Road, Slough, Bucks. (Tel.: Slough 24501.)

The resistance wire Evanohm, to which "Diallist" referred in September, is of American origin but is obtainable in this country from Gilby-Brunton, Ltd., 47, Whitehall, London, S.W.1.

A $3 \frac{1}{2}$-acre site has been chosen at Harlow, Essex, for the erection of a group of buildings to accommodate the recently-formed Siemens-Ediswan Research Laborator:es of which Dr. G. W. Sutton is director. Until the new buildings are occupied, laboratory space is being provided at the companies' works at Brimsdown, Woolwich and Blackheath.

Sound reinforcement equipment and multi-language interpretation facilities were provided by Tannoy for each of the 235 seats at the recent Suez Conference in Lancaster House, London.

The Ekco portable television receiver, which also incorporates v.h.f. sound, is being installed in the chauffeur-driven cars operated by Daimler Hire, Limited, London. Owing to the screening of the car's metal body and the limitation of space in the car, the receiver's telescopic built-in aerial has been replaced by one mounted externally. The screen can be seen only in the rear compartment.

Communal television aerial systems have now been installed in several Devon towns and at Sheringham, Norfolk, by J. S. Fielden, Limited. Work on other systems for the West Country, Lancashire and Cumberland is due to start this year.

Several radio applications of "Stick-a-seal" selfadhesive polyurethane foam are suggested by the suppliers, Sealdraught, Limited, of Chandos House, Buckingham Gate, London, S.W.1. Among them, loudspeaker mounting and sealing for glass panels of television sets. The material is available in three standard thicknesses (1/8th, $3 / 16$ th and $\frac{1}{4}$ inch) and a variety of widths up to 19 inches.

The production of Ardente hearing aids and miniature components has been transferred to the company's new factory at 8-12, Minerva Road, North Acton, London, N.W.10. (Tel.: Elgar 3923.)

Bel Sound Products Company, of Marlborough Yard, London, N.19, announce that they can supply p.t.f.e. machined from rod to any shape.

Welwyn Electrical Laboratories, Limited, component manufacturers of Bedlington, Northumberland, are now making thermistors.
The new offices and enlarged factory of Partridge Transformers, Limited, at Roebuck Road, Tolworth, Surrey, were recently opened by Mrs. V. R. Partridge, widow of the founder of the company.
Willesden Transformer Company, Limited, are now occupying a further 15,000 square feet at their new factory at Manor Park Road, Harlesden, London, N.W.10. (Tel.: Elgar 5445.)

Direct TV Replacements, of 134-136, Lewisham Way, New Cross, London, S.E.14, have been appointed distributors of Pinnacle valves.

Philips' new north-west regional headquarters at 20 , Cannon Street, Manchester, include a specially-designed demonstration room where it is planned to hold regular gramophone recitals.

## EXPORT NEWS

Communications equipment, including transmitters, receivers, aerials, power plant and ancillary gear, for Iran's Police Forces is to be installed by Redifon, Limited. The contract, valued at nearly $£ 500,000$, also calls for the setting up of a radio training school.

Loudspeaker Units.-The first consignment of Goodmans' recently-introduced "pressure" units-Trebax and Midax-was shipped to the United States at the end of August.

It was stated in a note on television receivers for Bangkok in this section last month that the $625-\mathrm{line}$ standard was employed. Although this was originally adopted Thailand now operates on American standards.

A report on the market for sound and television receivers in Italy has been prepared by the Export Services Branch of the Board of Trade. It concludes "If United Kingdom manufacturers wish to arrest the decline in their position and fight back against the commanding position gained by Germany in the last four years they will need to undertake aggressive sales publicity and offer, at competitive prices, sets that incorporate all the gadgets that have become virtually standard in Continental sets and pay particular attention to the rising popularity of f.m."
British manufacturers of domestic and marine radio equipment may be interested to know that Gough Industries Inc., a leading firm of wholesalers and distributors, of 819, East First Street, Los Angeles, has approached the British Consulate General with a proposal to help firms to sell their products in Southern California, Arizona, Nevada, Utah and Hawaii. The plan is that the company would receive samples and arrange for them to be exhibited to selected distributors within the Gough organization.
Electronic Equipment.-Feedback Control, Inc., 899, Main Street, Waltham, Mass., U.S.A., would like to represent United Kingdom manufacturers of electronic (other than communications) equipment who are not already represented in the U.S.A. They would act as representatives and/or a servicing organization throughout the United States.
Television Receivers.-Tebag AG., Lavaterstrasse 66 (Postfach), Zurich 27, Switzerland, are interested in representing United Kingdom manufacturers of television sets. The Swiss standards are 625 lines, negative vision modulation and f.m. sound.
Radiogramophones.-The Pentron Corporation, 777, S. Tripp Avenue, Chicago 24, Illinois, U.S.A., is interested in distributing good quality radiogramophones manufactured by British firms. They should cover the m.w. band as well as v.h.f./f.m.


CONTROL ROOM in the mobile studio recently brought into service by the B.B.C. for outside sound broadcosts. Provision is made for recording and reproducing programmes and sound effects, and a receiver is installed to pick up transmissions from commentators equipped with walkie-talkies. Two Mullard transmitter-receivers are provided for linking the studio to the nearest B.B.C. centre when Post Office lines are not avallable.

# Disc Recording Characteristics 

STANDARDIZATION AT LAST ?
SOME NOTES ON B.S. 1928 : 1955
By J. D. SMITH, B.Sc.

THERE has in the past been much controversy over the subject of recording characteristics. Numerous writers have quoted characteristics, often with considerable divergencies of opinion. Indeed the very mention of "recording characteristic" has been sufficient to unleash a spate of correspondence in the technical Press. In view of this it is very surprising that the revised British Standard 1928:1955*, issued over a year ago, has provoked almost no comment other than brief notices of its existence. Can it be that the new Standard settles once and for all every possible argument on the subject, or is the recording-characteristic-conscious section of the public largely unaware of its existence? Be this as

it may, it is perhaps worth while to examine this new characteristic, in view of the fact that the specification or its equivalent is now being adopted by many record manufacturers in this country and abroad.

Limitations of Standardization.-B.S. 1928 covers most aspects of recording and reproducing gramophone records and transcription recordings on discs. Speeds of rotation and various dimensional features of discs and reproducing equipment are specified. This much is relatively straightforward but the question of standardizing recording characteristics is very much more involved as the committee responsible for the Standard have been at pains to point out in an appendix.

The nature of these difficulties becomes apparent on examining Fig. 1, in which a complete recording and reproducing system is shown schematically. The studio equalizer is adjusted to compensate for studio and microphone deficiencies and to obtain the desired balance between high and low frequencies. The electrical signal at point $A$ is then such that when reproduced by means of a specified

[^3]monitor chain consisting of amplifier, loudspeaker and listening room, it has the balance and quality that the manufacturer desires: presumably a subjective judgment. To prevent adjacent grooves from overlapping at low frequencies, and to improve signal/noise ratio at high frequencies; this electrical signal is equalized to a known recording characteristic before being fed to the cutting head. During replay the output from the pickup is fed via an equalizer having a response which is the inverse of the recording characteristic, so that, save for any deficiencies in the system, the signal at B will be a replica of that at A. Then, if the same monitor chain as before were connected to $B$, the sound would be exactly as


Fig. 2. General form of standard recording characteristic.
the manufacturer intended. In practice the signal at $B$ is reproduced by a different system, usually incorporating yet another equalizer, the "tone controls," by which the listener introduces his personal preferences. The sound as finally reproduced may, therefore, differ from what the manufacturer had intended, tut it does so in a manner determined by the listener.

Now if a second manufacturer were to make a record of the same performance he would, in general, use a different monitor system and would equalize to produce a balance which he regarded as satisfactory. The signal at A would, therefore, differ from that produced by the first manufacturer at that point. The same is true of the replayed signal at

TABLE 1.
\(\left.$$
\begin{array}{|c|c|c|}\hline \hline \text { Time constant } & \text { Coarse groove } & \text { Fine groove } \\
\hline \text { Treble rise } t_{1} & \begin{array}{r}50 \mu \mathrm{sec} \\
450 \mu \mathrm{sec} \\
\text { Bass fall } \\
\text { Bass rise }\end{array}
$$ \& t_{2} <br>

\& t_{3} \& 3180 \mu \mathrm{sec}\end{array}\right)\)| 75 sec |
| ---: |
| $318 \mu \mathrm{sec}$ |

B, provided that matched recording and replay characteristics be used (though not necessarily similar to those used in the first case). This signal at B would have to be reproduced by the second manufacturer's monitor system in order to obtain the sound as intended by him.

Thus the whole picture becomes somewhat confused and all that standardization can do at present is to specify recording and replay characteristics which could be adopted by all manufacturers. This ensures that the listener can with certainty obtain at point B in his reproducing chain the electrical signal intended by the manufacturer. This the British Standard does and no more. Having done this there are still the differences between the

Fig. 3. Basic passive networks comprising a replay equalizer.


Fig. 4. Equalizer employing negative feedback. The resistor $R_{5}$ (shown dotted) is normally omitted, but may be included if necessary to limit the low-frequency boost.
various manufacturers' monitor systems: as pointed out in the Standard, it would be highly desirable to standardize these but at present this is impractic-

| FUNCTION | NETWORK | TRANSFER FUNCTION |
| :---: | :---: | :---: |
|  |  | $\frac{V_{0}}{V}=\frac{1}{\sqrt{1+4 \pi^{2} t^{2} t_{1}^{2}}}$ |
| BASS RISE |  | $\frac{V_{0}}{i R_{2}}=\sqrt{1+\frac{1}{4 \pi^{2} t^{2} t_{2}^{2}}}$ |
|  |  | $\frac{V_{0}}{V}=\frac{1}{\sqrt{1+\frac{1}{4 \pi^{2} t^{2} t_{3}^{2}}}}$ |
|  | OVERALL $\sqrt{\frac{i t}{\left[1+4 \pi^{2} t^{2} t\right.}}$ | ansfer function $\frac{\frac{1}{\pi^{2} t^{2} t_{2}^{2}}}{\left[1+\frac{1}{4 \pi^{2} t^{2} t_{3}^{2}}\right]}$ |
| $\begin{gathered} n=\frac{t_{3}-t_{2}}{t_{2}} \\ R C_{4}=t_{2} \\ R C_{B}=\frac{t_{1} t_{3}}{t_{3}-t_{2}} \end{gathered}$ |  | TRANSFER FUNCTION AS FOR CASCADED NETWORKS | able. The manufacturer's preferences in the matter of balance and so forth must be regarded as part of the actual performance and as such may not be subjected to standardization; the same is true of adjustments made by the listener.

The above discussion seems to suggest that there are still many loopholes in the Standard. There is perhaps some truth in this but nevertheless it is a great step forward to have two characteristics, one for "coarse groove" and one for " fine groove" recordings, clearly and simply defined so that the listener is no longer at the mercy of opinion in this matter. It is certainly to be hoped that all manufacturers will adopt them.
The New Standards.
The new standard characteristics are very conveniently defined in terms of the timeconstants of equalizing networks. Fig. 2 shows diagrammatically a recording characteristic. There are three portions to this curve: at A there is a treble rise defined by time-constant, $t_{1}$, so that at high frequencies the curve rises at a rate of 6 dB per octave. At B


Fig. 5. Equivalent circuits of Fig. 4 at (a) low, (b) mid-band, and (c) high frequencies.


Fig. 6. Complete equalizer for B.S.S. 1928 : 1955 characteristics. VI may be Z729, EF86, 6BR7, etc. Mid-band gain approx. 10. Switch positions: I-B.S.S. Coarse Groove, 2-B.S.S. Fine Groove.
there is a bass fall and a second time-constant, $t_{2}$, defines this. At low frequencies the response does not fall away indefinitely because a bass rise of time-constant, $t_{3}$, is included. Table 1 gives the values of the time-constants as set out in the Standard.

Replay Equalizers.-A replay equalizer could be constructed using passive networks chosen to give the appropriate time-constants, remembering that a rise in recording characteristic must be matched by a fall in replay characteristic. The three networks must be cascaded in such a manner that they do not interact one with another; alternatively a single passive network incorporating all the necessary time-constants may be used. Such networks are shown in Fig. 3.

It is preferable, however, to use a valve with selective feedback to provide equalization and to incorporate the time-constants in the feedback loop. Fig. 4 shows such a circuit. In Fig. 5 are shown the three circuit conditions at low, mid-band and high frequencies. Notice that, as Fig. 5(a) shows, the gain at low frequencies, where maximum boosting is required, is limited to that available from the velie. In this wav the required bass fall is provided without actually including a further time-constant: the ratio of the effective time-constant of this bass
fall to that of the bass rise is the ratio of the maximum gain of the stage to that at mid-band, where a moderate amount of feedback is applied as shown in Fig. 5(b). If in a particular circuit this ratio is too great, $\mathbf{R}_{5}$ may be included to provide a small amount of feedback at low frequencies. In a similar way there is a limitation of the high-frequency attenuation when the condition of Fig. 5(c) is reached, the gain then being unity (if $R_{1}=R_{2}$ ). However, this undesired limitation is not serious in a properly designed circuit. The actual timeconstants are given by $C_{1}\left(R_{5}+R_{4}\right)$ for bass rise and $C_{2} R_{4}$ for treble fall.

Fig. 6 shows a circuit with suitable component values. Note that the switch may have as many positions as desired so as to incorporate equalization for older recordings. The load on this stage should not be heavier than 1 megohm or the available gain will be reduced and full bass boost will not be provided. If the stage must be more severely loaded it is possible by reducing $\mathbf{R}_{2}$ to obtain the necessary bass boost at the expense of overall gain.

## DO YOU KNOW?

THE length of the dipole for a Band II aerial? The relationship between m.k.s. and c.g.s. units? The address of the International Amateur Radio Union?

The base connections for a LN309 valve?
What external resistance is needed in series with a 25 -volt meter ( $1,000 \mathrm{ohms} / \mathrm{V}$ ) to read voltages up to 500?

If a licence is required to operate a transmitter for the control of a model?

The answers to these and innumerable other technical and organizational questions can be found in the 1957 Wireless World Diary-the vade mecum of all who have an interest in radio.

The Diary, now in its thirty-ninth year of publication, includes, in addition to the usual week-at-an-opening diary pages, an eighty-page reference section. It is obtainable from booksellers and newsagents, price 6 s (leather) and 4 s 3 d (Rexine) including purchase tax. Overseas prices are, respectively, 5 s and 3 s 6 d , plus 2 d postage.
"Full-Range Electrostatic Loudspeakers."-The third line from the bottom of column 2 of p. 486 of the October issue should read " $. . . \mathrm{C}=$ equivalent capacitance $=4 t_{\mathrm{r}}{ }^{2} / \mathrm{K}$, where $\mathrm{K}=$ motional stiffness. . . ."

# Transistor D.C. Amplifier 

LOW-NOISE CIRCUIT FOR
MILLIMICROAMPERE SIGNALS
By D. M. NEALEネ and FRANCIS OAKES $\dagger$


#### Abstract

An amplifier is described which provides a current gain of 1,000 ; power gain, 45 dB : zero stablity, $0.001 \mu \mathrm{~A}(10 \mu \mathrm{~V})$; and a frequency response extending beyond $20 \mathrm{kc} / \mathrm{s}$. A push-pull grounded-emitter stage is followed by a push-pull grounded-collector stage. A fifth transistor in a negative feedback loop limits the effects of collector leakage current variations, and by stabilizing the first-stage collector voltage restricts the effect of transistor noise. The amplifier was developed for use with barrier-layer photocells at very low light intensities. It has excellent linearity and can also be used in conjunction with a semiconductor diode to provide a high. impedance low-level r.f. voltmeter usable up to frequencies limited only by the performance of the diode.


FOR the amplification of small direct currents, a battery-fed transistor amplifier offers several advantages. Warm-up time is greatly reduced and problems of supply voltage stabilization are virtually eliminated. The effects of low-frequency noise ${ }^{1}$ and the rapid rise of leakage current with rise in temperature can be controlled by careful design so that, where a medium input impedance is required, the transistor-operated amplifier provides a better performance than a thermionic amplifier.

Whereas it is possible to eliminate the effects of temperature by carefully matching transistor characteristics and also by applying negative feedback, there seems to be no way of countering the effects of noise. The simple circuit of Fig. 1 was therefore used to find how small an input current could be definitely distinguished from noise. It was soon found that, as has been reported before ${ }^{2}$, there is a marked increase in semiconductor noise at higher collector voltages. A collector voltage of at least 0.2 V is of course required to make the transistor operate in the high-alpha region, but it seems that the noise remains substantially constant until the collector voltage exceeds 1.0 V .

With collector voltages between these two limits, the noise fluctuations were observed when a sensitive galvanometer was used as a balance indicator. With Mullard OC71 transistors, fluctuations due to noise were equivalent to about $0.001 \mu \mathrm{~A}$ at the input. Since a comparable stability is not readily produced in a thermionic amplifier of comparable input impedance ( 5 to $10 \mathrm{k} \Omega$ ), an effort has been made to reduce to an insignificant level the drift due to other causes.

The most difficult part of the design appeared to be that of keeping the collector-emitter voltage, $\mathrm{V}_{\mathrm{c}}$, in the range $0.2-1.0 \mathrm{~V}$ over a reasonable range of ambient temperatures. Using a push-pull groundedemitter input stage and collector load resistances of the order of $20 \mathrm{k} \Omega$, changes in collector leakage current limited the range of satisfactory operation without stabilization to about $20^{\circ} \mathrm{F}$, e.g. $60^{\circ}-80^{\circ} \mathrm{F}$.

The standing current in a transistor stage may be stabilized ${ }^{3}$ by the insertion of a resistance in the emitter circuit. The resistance provides negative feedback restricting the variations in the d.c. component of the collector current. Its effect on the a.c. component is usually minimized by the provision of
a bypass condenser. In a single-ended d.c. stage, this type of stabilization offers no advantage, since drift and gain are reduced in the same proportion. In a push-pull stage, however, a common emitter resistance provides stabilization against the effects of in-phase current changes due to leakage current variations, whilst making no reduction in the gain so far as push-pull signals are concerned.

For the type of circuit shown in Fig. 2 it can be shown ${ }^{3}$ that the stability factor S is given by

$$
\begin{aligned}
\mathrm{S} & =\frac{\mathrm{dI}_{c}}{\mathrm{dI}} \\
& =\frac{1+\mathrm{R}_{\mathrm{B}} / \mathbf{R}_{\mathrm{B}}}{1+(1-\alpha) \mathbf{R}_{\mathrm{B}} / \mathbf{R}_{\mathrm{E}}}
\end{aligned}
$$

In order to keep $\mathrm{V}_{c}$ substantially constant, a low

* Ilford, Ltd.
+ Ferguson Radio Corporation


Abcve: Fig. 1. Simple push-pull amplifier circuit

Right: Fig. 2. Stobilized single-ended amplifier.



Fig. 3. Addition of a transistor ( $T_{5}$ ) to provide amplified feedback.

Below: Fig. 4. Modification of circuit of Fig. 3 to eliminate the separate stabilizing battery.

value of $S$ is required. The above equation shows that $S$ falls to unity when $R_{B} \ll R_{E}$. Since $R_{B}$ must be kept high in order to avoid unnecessary shunting of the input signal, however, a low value of S necessitates a high value of $\mathrm{R}_{\mathrm{E}}$. As the collector leakage current passes also through $R_{E}$, the voltage drop across $\mathbf{R}_{\mathrm{E}}$ reduces $\mathrm{V}_{\mathrm{c}}$ and largely neutralizes the stabilizing action provided.

A more useful criterion for design is the collectoremitter slope impedance, $\mathrm{dV}_{c} / \mathrm{dI}_{c 0}$. Assuming limiting values of $\mathrm{V}_{c}$ and $\mathrm{I}_{c o}$ of 0.8 V and $16 \mu \mathrm{~A}$ respectively, the highest acceptable value of $-\mathrm{dV}_{\mathrm{c}} /$ $\mathrm{dI}_{\text {co }}$ is $50 \mathrm{k} \Omega$. The value realized in a given circuit is given, using conventional symbols, by

$$
\begin{aligned}
\frac{\mathrm{dV}_{c}}{\mathrm{dI}_{c o}} & =\frac{\mathrm{dI}_{c}}{\mathrm{dI}_{c o}} \cdot \frac{\mathrm{dV}_{c}}{\mathrm{dI}_{c}} \\
& =-\mathrm{S}\left(\frac{\mathrm{R}_{\mathrm{E}}}{\alpha}+\mathrm{R}_{\mathrm{I}}\right)+\frac{\mathbf{R}_{\mathrm{E}}}{\alpha}
\end{aligned}
$$

If $R_{F} \gg R_{B}$ and $R_{E} \geqslant R_{L}$,

$$
S \approx\left(1+R_{B} / R_{E}\right)
$$

and $\frac{\mathrm{dV}_{\epsilon}}{\mathrm{dI}_{c o}} \approx-\left(1+\frac{\mathrm{R}_{\mathrm{B}}}{\mathrm{R}_{\mathrm{E}}}\right)\left(\mathrm{R}_{\mathrm{E}}+\mathrm{R}_{\mathrm{I}}\right)+\frac{\mathrm{R}_{\mathrm{E}}}{\alpha}$

$$
\approx \frac{\mathbf{R}_{\mathrm{E}}}{\alpha^{\prime}}-\left(\mathrm{R}_{\mathrm{B}}+\mathbf{R}_{\mathrm{L}}\right)
$$

Hence, to obtain low values of $d V_{c} / \mathrm{dI}_{c o}$ with reasonable values of $R_{L}$, it is necessary to put $R_{B} \ngtr 3 R_{L}$ and $R_{E} \approx \alpha^{\prime} R_{L}$. This high value of $R_{E}$ results in an inconveniently high battery voltage which must, to provide a satisfactory value of $\mathrm{V}_{c}$, have high stability.

Although simple forms of negative feedback are of little use when the collector voltage must be stabilized within narrow limits, good results may be obtained
by using an additional transistor to apply amplified feedback.

Fig. 3 is the schematic circuit of an arrangement of this type. $\mathrm{T}_{1}$ represents the transistor to be stabilized, $\mathrm{T}_{5}$ the transistor used for applying an amplified control signal to the base resistor, $R_{B}$, of $T_{1}$. If the collector current of $T_{1}$ increases, the reduction in collector-emitter voltage of $T_{1}$ is transferred by the feedback battery, $\mathrm{V}_{f}$, to increase the emitter-base potential of $T_{5}$. Any such increase in emitter-base potential results in a large increase in base current in $T_{5}$, limited only by the internal base and emitter resistances. The increased base current in $\mathrm{T}_{5}$ in turn produces a large increase in the collector current of $T_{5}$, so increasing the voltage developed across $R_{7}$ and reducing that across $R_{B}$. In consequence, the base current of $\mathrm{T}_{1}$ is substantially reduced, enabling the collector current through $\mathrm{R}_{1}$ to be maintained practically constant. It is fortunate that the effect of a temperature rise of $T_{5}$ serves to assist the temperature compensation applied to $\mathrm{T}_{1}$. Since the emitter-base potential of $\mathrm{T}_{5}$ is very small so long as a positive base current is flowing, the collector voltage of $T_{1}$ is stabilized at the voltage of the battery $V_{f}$.

The additional battery, $V_{f}$, was eliminated in the final amplifier by modifying the circuit of Fig. 3 to that shown in Fig. 4. The transistor $\mathrm{T}_{3}$ represents the grounded-collector stage which follows the grounded-emitter transistor $\mathrm{T}_{1}$. As the emitterbase potential of $T_{3}$ is very small, the potential of the emitter of $\mathrm{T}_{3}$ is effectively the same as that of the collector of $\mathrm{T}_{1}$. If a relatively high battery voltage is used, e.g. 13.5 volts, 1 volt may be dropped across $\mathbf{R}_{4}$, whilst at least $90 \%$ of the collector excursion of $T_{1}$ is still developed across $R_{6}$. The circuit therefore works in the same way as Fig. 3, stabilizing the collector-emitter voltage of $\mathrm{T}_{1}$ to the voltage dropped across $\mathrm{R}_{4}$.

The circuits shown in Figs. 3 and 4 provide a high degree of negative feedback not only for variations of collector current due to changes in the collector leakage current, $I_{\text {co }}$, but also for changes in collector current due to the amplification of intentionally introduced signal currents fed in at the base of $T_{1}$. In the complete push-pull amplifier, however, the emitter circuit resistors $\mathbf{R}_{3}$ and $\mathbf{R}_{6}$ are common to both transistors in each stage It follows, therefore, that push-pull signal currents will cancel in these resistors, whereas the effect of leakage currents will be additive.

A circuit developed on these principles is shown in Fig. 5. It will be seen that a single transistor $\mathrm{T}_{2}$ is used to control the standing base currents of the transistors $T_{1}$ and $T_{2}$ in the first stage. At any given temperature the output indication for zero input current may be made independent of source impedance by a simple setting-up procedure. The input terminals are first short-circuited and $P_{z}$ is adjusted to set the output meter to zero. The input terminals are then open-circuited and $P_{c}$ is adjusted to correct any consequent shift in the zero. Provided the characteristics of $\mathrm{T}_{1}$ and $\mathrm{T}_{2}$ are reasonably well matched, the setting of $P_{c}$ produces negligible effect on the zero for short-circuited input.

Algebraic analysis of the final circuit leads to cumbersome expressions which cannot be simplified without sacrificing accuracy. The slope resistance $\mathrm{dV} / \mathrm{dI}_{c o}$ was therefore established experimentally,
and found to be of the order of $2 \mathrm{k} \Omega$. This represents an improvement by a factor of 15 over the circuit of Fig. 2.

It was found that the short-circuit zero was commendably stable over the temperature range $50^{\circ}$ to $85^{\circ} \mathrm{F}$. The open-circuit zero varied considerably however, because the leakage current, internal base- and emitter-resistances of the transistors $\mathrm{T}_{1}$ and $\mathrm{T}_{2}$ vary with temperature in such a way that a setting of the potentiometer $P_{c}$ which is satisfactory at one temperature is not satisfactory at another. Some means was sought whereby the open-circuit and short-circuit zero could be made coincident at two points in the ambient temperature range, with a reasonable agreement in between.

Reversal of the sense of $P_{c}$ occurs when the collector-base voltage of $T_{5}$ falls sufficiently to equal the base-emitter voltages of $T_{1}$ and $T_{2}$, Under these conditions the base currents of $\mathrm{T}_{1}$ and $\mathrm{T}_{2}$ become zero and their collector currents are given by

$$
I_{c}=\left(1+\alpha^{\prime}\right) I_{c o}
$$

With representative values of $\alpha^{\prime}, I_{c o}$, and internal base and emitter resistances, it may be deduced that the base-emitter voltage amounts to some

hundred millivolts. At temperatures higher than that of this special case, the collector-base voltage of $\mathrm{T}_{5}$ will become smaller than the base-emitter voltages of $T_{1}$ and $T_{2}$. The base currents of $T_{1}$ and $T_{2}$ are then reversed and it is therefore to be expected that the sense of $\mathrm{P}_{c}$ will be reversed also; whereas a reduction of base resistance of $T_{1}$ increases both base and collector currents at moderate ambient temperatures, at higher temperatures it increases the negative base current and so reduces the collector current.

In the final amplifier circuit, shown in Fig. 6, this reversal of sense of $P_{c}$ is used to provide compensation for the effects of temperature on the open-circuit zero. A further potentiometer $\mathrm{P}_{b}$ is used, the sense of which is not reversed by temperature effects. The slider of this control is returned to the emitter of $\mathrm{T}_{5}$ which is at a potential a few millivolts positive relative to the emitters of $\mathrm{T}_{1}$ and $\mathrm{T}_{2}$. The addition of $\mathrm{P}_{b}$ makes it possible to use a setting of $\mathrm{P}_{c}$ which provides temperature compensation and which at the same time allows the short-circuit and open-circuit zeros to be made coincident. This is done without seriously reducing the temperature stabilization provided by $\mathrm{T}_{5}$.

The optimum setting of $\mathbf{P}_{\mathrm{c}}$ is deduced by plotting (with open-circuit input) the meter current versus ambient temperature for various combinations of settings of $P_{c}$ and $\mathrm{P}_{z}$. Coincidence of opencircuit and short-circuit zeros is effected by adjusting $P_{b}$ according to the procedure described above in connection with Fig. 5.
When the three controls $\mathrm{P}_{b}, \mathrm{P}_{c}$ and $\mathrm{P}_{z}$ are adjusted to their optimum positions, very fair temperature compensation is provided over the ambient temperature range $55^{\circ}$ to $85^{\circ} \mathrm{F}$ and, as shown in Fig. 7, a reasonable coincidence is main-

Above: Fig. 5. Amplifier with zero adjustments to make output for zero input current independent of source of impedance.

Fig. 6. Final circuit with additional compensation for the effect of temperature on the zero setting.


'Fig. 7. Measured variation of zero setting with temperature in the circuit of Fig. 6.
tained over this range between the open-circuit and short-circuit zeros. At its worst, the zero drift in this range amounts to about 10 millimicroamps at the input.

The amplifier was, however, required for use with apparatus demanding a short-term stability of $1 \mathrm{~m} \mu \mathrm{~A}$. This was attained by mounting the transistors in holes drilled in a block of brass and subsequently wrapping the whole amplifier in several layers of foamed rubber. These measures ensure that the temperatures of the transistors are identical and that their temperature drift is very slow. If a correspondingly good long-term stability had been required, it could have been obtained by using a simple bimetallic thermostat to maintain the outer case of the amplifier at a substantially constant temperature. To eliminate initial drifts due to internal dissipation in the transistors, the amplifier is left in operation continuously. The -total current consumption is less than 3 mA and a battery life of several months is obtained from three 3-cell torch batteries.

The amplifier gain has been measured over a useful range of ambient temperatures. It will be seen from Fig. 9 that a substantially uniform gain is maintained from $37^{\circ} \mathrm{F}$ (the lowest temperature at which tests were made) to $.93^{\circ} \mathrm{F}$. The upper limit is set by the bottoming of $\mathrm{T}_{5}$.


Fig. 8. Power gain variation with temperature in the circuit of Fig. 6 .

Simple tests show no phase shift or irregularity in the frequency response of the amplifier below the limit imposed by the alpha cut-off frequency. The main limitation on the value of the amplifier arises from the low-frequency noise, corresponding to about $0.001 \mu \mathrm{~A}$ at the input.

Experimentally the Mullard Type OC71 transistors used for the first stage of the amplifier were replaced by Type OC70 transistors. The manufacturers' literature indicates that the noise level of the OC70 is some 6 dB lower than that of the OC71. Consequently the unsteadiness of the zero might have been expected to be halved by using the low-noise type. In practice, however, no improvement could be discerned. The OC70 provides a slightly lower gain and it is more difficult to select matched pairs. Consequently Type OC71 transistors were retained throughout.

The transistors used in each stage were selected so that, as nearly as possible, those in each pair had identical values of $x^{\prime}$ and $I_{c o}$. The pair having the lower value of $I_{c o}$ was, of course, used for the input stage. Although only a limited number of transistors were available at that time, it was found fairly easy to match values of $\alpha^{\prime}$ to about $3 \%$ and $I_{c o}$ to about $10 \%$.

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## Mohile Sound and TV Lahoratory



TO facilitate the testing of prototype sound and television receivers in different ports of the country Cossor's have brought into use a mobile radio loborotory. The vehicle, which is equipped with its own power supply, carries an extendible oerial system remotely controlled from the test bench. The equipment includes Model 1322 "Telecheck " and marker generator for Bands I and III, an f.m. receiver alignment generator ond oscillographs.

# AUTOMATIC WAVEMETER Chlibration 

THE precision of wavemeter measurement cannot exceed the accuracy of the calibration, which is commonly recorded in a handbook or a calibration chart attached to the instrument. Re-calibration is necessary at intervals dependent on the use made of the wavemeter. Where the instrument is subjected to rigorous working conditions, as experienced with the Armed Services, the necessity of frequent recalibration becomes onerous.
The Type BC221 heterodyne frequency meter, for instance, occupies a skilled technician for fifteen days, for it has 3,252 calibration points. The wavemeter in question is one used in large quantities and in order to speed up re-calibration an automatic apparatus $\dagger$ has been designed. The whole process of re-calibration from the generation of standard frequencies to the photographic recording of dial readings is entirely automatic and the apparatus can deal with five wavemeters per day.
A series of calibration frequencies (of the order of one part in $10^{7}$ accuracy) are provided by a frequency standard. These are compared with the output frequency of the wavemeter under test. The wavemeter output frequency is continuously varied by a mechanical drive which simultaneously operates a mechanical counter. There is always a direct relationship between the counter and the position of the wavemeter tuning control.
The calibration frequencies and the continuously changing output frequency of the wavemeter are compared, using the heterodyne principle. After comparison an output is obtained which causes

By D. P. THURNELL,* B.5c. (Eng).

pulses are applied as one input to a mixer circuit (4).
The output from the wavemeter (5) under test passes to an amplifier (6) and thence to the mixer (4). The wavemeter has its tuning dial firmly coupled to a mechanical drive consisting of an electric motor (7) and a precision gear box (8).

The output from the mixer circuit is amplified by a low-frequency amplifier (9), and applied as the input to a mono-stable trigger circuit (10).

One pulse output from the trigger circuit is fed via a commutator device (11) to effect the triggering of a thyratron circuit (12) which provides a firing pulse for the photographic flash tube (13) arranged evenly to illuminate a mechanical multi-digit counter (14). This is driven at a suitable speed to synchronize with the tuning control of the wavemeter and indicates the same figures as the wavemeter dial reading. Alternatively, it can be arranged to indicate the angular position of the wavemeter dial. When used with the BC221 heterodyne frequency meter the former arrangement is employed.

In correct alignment with the viewing aperture of the counter is a camera (15). At each flash of the tube (13) an image of the counter is recorded. The camera has a film wind-on control which is mechanically coupled to a solenoid-actuated mechanism (16).

* G. and E. Bradley, Ltd., Wembley.

Fig. I. Block diagram showing main components and functions of the equipment.
indications of the counter to be automatically recorded upon the occurrence of each zero-beat frequency.

Fig. 1 is a block diagram showing the technique employed. The output from the crystal-controlled standard frequency oscillator (1) is fed to an electronic frequency divider (2). The output obtained from the divider is at a frequency equal to the required increment of frequency between successive calibration points. This output is applied to a combined squaring and pulse-generating circuit (3) which provides at its output a series of squared waveform pulses at a recurrence frequency equal to the chosen increment of frequency. Each pulse is of a suitable amplitude and of a sufficiently short time duration to ensure a spectrum of substantially equal amplitude harmonics up to the upper limit of the frequency range of the instrument undergoing calibration. The

[^4]

The film control mechanism is supplied with energizing current pulses from a thyratron and relay circuit (17) whose control input is connected to an output from the trigger circuit (10) to ensure that the film is moved on for each exposure.

It will be assumed that the wavemeter has a frequency range of 100 to $200 \mathrm{kc} / \mathrm{s}$, that its tuning control is capable of being read to one part in 10,000 of its complete range of movement, and that calibration is to be effected at every $2 \mathrm{kc} / \mathrm{s}$ over the wavemeter frequency range. In such circumstances the desired recurrence frequency for the pulse output from the circuit (3) is $2 \mathrm{kc} / \mathrm{s}$ and the standard frequency source (1) may operate at, say, $100 \mathrm{kc} / \mathrm{s}$ with the divider (2) effecting division by a ratio of $50: 1$. The motor (7) will be arranged to cause movement of the control of the wavemeter (5) over its complete range of movement while counter (14) is being moved from zero to a count state of 10,000 .

The pulse input to the mixer (4) from circuit (3) comprises a series of equal-amplitude harmonics of the recurrence frequency (i.e., $2,4,6,8 \mathrm{kc} / \mathrm{s}$ and so on upwards) to a frequency above the maximum operating frequency ( $200 \mathrm{kc} / \mathrm{s}$ ) of the wavemeter. This implies a pulse width less than $1 \mu \mathrm{sec}$. In consequence there will be produced at the output of the mixer circuit (4) a succession of zero beats, the first occurring when the wavemeter output is $100 \mathrm{kc} / \mathrm{s}$, the second when the output is $102 \mathrm{kc} / \mathrm{s}$, the third when the output is $104 \mathrm{kc} / \mathrm{s}$ and so on. The other harmonics of the $2-\mathrm{kc} / \mathrm{s}$ pulse repetition frequency are filtered from the mixer output in the amplifier (9) in which the bandwidth is arranged to be of very limited and low value, say $50 \mathrm{c} / \mathrm{s}$ in the present instance, so that a 1rigger input will be supplied by the amplifier to the trigger circuit (10) just before, but sufficiently close to, each occurrence of a zero beat output from the mixer. The small error so introduced (in this case about $0.05 \%$ ) can be compensated for, if the wavemeter-dial rotation/ frequency law is substantially linear, by offsetting the counter indication by a suitable amount from the dial indication. If this is not possible, iticreased accuracy can be obtained at the expense of speed by reducing the bandwidth of the amplifier (9).

When triggering of the circuit (10) occurs, an output of appropriate polarity is available to operate the commutator circuit (11). This commutator is necessary in order to delay the photographic exposure until the next cccasion on which the

Plan view of chassis. The dotted lines indicate subdivisions of function.
counter figures are all in line, since the least significant digit drum is continuously in motion. This is achieved by switching a bi-stable (flip-flop) circuit " on " with the pulse from the trigger circuit (10) and "off" with the next pulse derived from a photocell pick-up head energized by light interrupted by a slotted disc. This disc has ten slots and is fixed to the counter input shaft, the position of each slot corresponding to one of the ten positions in which all figures are in line. To avoid a blurred image a flash duration of approximately $100 \mu \mathrm{sec}$ is employed. When this occurs a photographic recording is made of the instantaneous indication of the counter.

The output pulse from the trigger circuit (10) terminates after a pre-determined time interval which is made long enough (with respect to the speed of movement of the wavemeter control) to prevent another triggering of the circuit (10) until the next zero beat is approached. Fig. 2 shows the amplitude/time output of the amplifier (9) during operation The thyratron and relay circuit (17) operates after the exposure. By closing the associated relay contacts the solenoid-actuated mechanism (16) winds on the camera film to the next recording position in readiness for the next cycle of operation which occurs when the next zero beat is reached.
This procedure is repeated at wavemeter output frequency intervals equal to the recurrence frequency (i.e. $2 \mathrm{kc} / \mathrm{s}$ ) of the pulses fed to the mixer over the whole of the frequency range for which calibration of the wavemeter is desired.

The characteristics of the optical system are so chosen that the developed film is produced at the correct size and pitch of figures and with a positive


Part of the automatic calibrctiln equipment showing mechanical drive for wavemeter tuning control.
image (black on white) so that it may be cut and mounted directly into a calibration booklet for the particular instrument, in which columns of the appropriate frequencies have already been printed.

The range of instruments which may be calibrated by theapparatus described can be extended by the employment of ancillary equipment, such as a frequency divider interposed between the wavemeter and the r.f. amplifier, when the wavemeter is one operating al very high frequencies. It may also be more convenient to use a servo-
 mechanism for the drive system between the motor (7) and the elements (5), (11) and (14).

Operation of the apparatus over other frequency ranges can best be described by examples. For instance, calibration has been effected at $10-\mathrm{kc} / \mathrm{s}$ intervals from 20 to $40 \mathrm{Mc} / \mathrm{s}$. This required the


Fig. 2. Ou:put amplitude/time characteristic of amplifier (9) during operation.


Specimen of calibration table as it appears af er the film record has been registered with the printed frequency table.
generation of a pulse of duration less than $10 \mu \mathrm{sec}$ with a time jitter less than $1 \mu \mathrm{sec}$.

In general, almost any frequency-calibration problem can be tackled by similar equipment. An important application is the automatic calibration of radio receivers where a high order of accuracy and a large number of calibration points are involved. The calibration record may then be in the form of a flexible scale (e.g., $35-\mathrm{mm}$ film), coupled to the tuning mechanism, on which are recorded the frequencies to which the receiver may be tuned.

## Further Developments

In order to eliminate the complication involved in photographic processing, a printing counter is under development. This is designed to replace the optical system and is provided with a mechanical storage device which enables a print to be made without interrupting the continuous drive to the input shaft.
The advantages of the system described may be summed up as follows:
(1) The elimination of human error during calibration;
(2) The elimination of human error in copying figures or in interpolation;
(3) The reduction of electrical error due to longterm temperature changes, thus disposing of the necessity for using a temperature-controlled calibration room;
(4) The increased speed of output and consequent reduction of staff.

As in other fields to which automation can be introduced, these advantages are economically realized where adequate numbers of similar instruments are to be dealt with and where the requirements of accuracy and multiplicity of calibration points are such as to justify the design of suitable equipment. Where large numbers of instruments are involved, a considerable economic advantage may be gained over hand-calibration methods.

## ILEIUNERS TO THE EIDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

## Television Systems

IN your July issue you again return to the question of choice between a 625 -line television system and a $405-$ line system for colour TV. Although I put my arguments at some length in my previous letter, there is one point in your Editorial to which I must reply.

You assert that any county sufficiently developed industrially to have any kind of TV service would probably make is own receivers. This is indeed a most defeatist attitude to adopt in view of the vital importance of our own export market. It will be many years before some of the underdeveloped countries can possibly start making television receivers, especially those for colour reception, though popular demand may well bring a television service to them. Are we then to sit back and see our competitors reap a rich harvest by exporting to these countries while we must wait to see whether or not factories are erected to make receivers on the spot? You may be sure that the introduction of television in a country will not await the indigenous production of receivers.

I am told that in Iraq, for instance, about 100 people watch each set whenever programmes are transmitted. This is surely a fair indication of the popularity of television and of future demand for imported sets. A radio set is much easier to make than a television set, yet in 1955 Germany sold some twelve million pounds worth of radio sets overseas while Britain also sold about three and a half million pounds worth in the same period. Surely, if it is possible to export these large quantities of radio sets, then there can be no possible justification for assuming that television sets will not be a profitable export.

As I said before, a healthy export market must be based on home demand. Colour television is certainly on its way and the great potential markets which will arise in these, at present, underdeveloped countries will undoubtedly be for sets operating on the 625-line system. All the more reason why we in this country should not only fall into line but lead our competitors in this field.

London, W.1.
ROBERT RENWICK.
WITH reference to your article "Television Picture Quality" in your August issue, reprinted from Wireless Engineer, May I ask why W. T. C., in studying the vertical definition does not mention the 0.7 coefficient, which is normally accepted, due to the fact that not all the active lines are capable of representing separate resolution lines?

Madrid.
MARIANO MATAIX.

## Comment:

THE coefficient of 0.7 does not in any way affect comparisons of the vertical definitions obtainable with different television systems because it affects them all equally. It does, however, affect a comparison of vertical and horizontal definition, in that when it is included the nominal figure for the relative definition in the two directions is altered.

The mechanisms limiting definition in the two directions are quite different both physically and in their visual effect. Any method of comparison is thus largely arbitrary. If I had said that the vertical definition is to be considered as equal to 0.7 of the number of active lines some of my statements about the relative definitions in the two directions across the picture would have been modified, but the comparisons of one system with another would have been unaffected.

The experimental fact that a 405 -line picture with unrestricted bandwidth is more pleasing than a 625 -line picture with $5 \mathrm{Mc} / \mathrm{s}$ bandwidth is not affected by whether
one says that the normal 405 -line system has roughly equal horizontal and vertical definition, as I did, or whether one says that it has lower vertical than horizontal definition, as I should have done had I introduced the 0.7 coefficient.
W. T. C.

I HAVE been much interested in the controversy on colour standards for television. The need for some standards to be laid down if we are to make any practical advance towards obtaining a colour service is obvious, and I should have thought that the unknowns are at present too numerous for any reasonable estimate of the economics of receiver production, let alone the manufacture of equipment for programme transmission.

It would be interesting to know what the official Post Office attitude is towards colour. The main trunk lines for both B.B.C. and I.T.A. networks are limited by design to a $3-\mathrm{Mc} / \mathrm{s}$ bandwidth, and to convert them to a wider bandwidth would take time and is probably out of the question while the I.T.A. is still expanding. Similar considerations apply to the local and temporary circuits, due to the P.O. amplifier design. Colour in less than five years would therefore seem to be geared to a $3-\mathrm{Mc} / \mathrm{s}$ bandwidth and existing communications.

I think it important not to overlook the communication aspect of this problem. Programme operation entails remote control of the transmitter, and transmitter and receiver design are therefore only part of the problem.

Among other parameters which have been little mentioned but whose standardization is probably essential before a colour service can be considered seriously are the transfer characteristic of the receiving tube, and the stray light factor at the receiver. In the U.S.A. this transfer exponent is standardized at 2.2. Is this a correct standard? Again, would 3 per cent of maximum luminance be regarded as a reasonable stray light factor? Unless they are given these two parameters, it is difficult to see on what basis the designers of commercial camera chain equipment for colour would work.
Woking.
R. F. COLVILE.

## Noise in Carbon Resistors

I AM indebted to Dr. D. A. Bell for his comments on my article on "Characteristics of Fixed Resistors". Regarding the point of current noise, I have endeavoured to be cautious when making statements concerning resistors generally, as I would not like to state that all resistors have the same characteristics with regard to noise until all types of resistor have been measured. Chipman, in addition to Hollins and Templeton, has shown the low-frequency power spectrum to be true for certain types of resistor, and as more data is obtained on, for instance, metal film and oxide film resistors the figure "below about $10 \mathrm{c} / \mathrm{s}$ " should certainly be changed to "below about $1 \mathrm{c} / \mathrm{s}$ ".

The reference to $\log f_{2} / f_{1}$ is correct, and if my book is referred to (on page 49) it will be seen to be correct there; the omission occurred in the process of transcription from the book.

Whilst the figures given in Table 3 are acceptance limits they are, in fact, based on the results of a large number of measurements and they do represent what is implied-the maximum noise one would normally expect on typical resistors. The formula $0.5 \mu \mathrm{~V} / \mathrm{V}$ is still used in RCS 112 for cracked carbon resistors but I would agree that this does not purport to reflect the true noise level. In the absence of precise data it serves as a good guide to the design engineer and it is clear from Table 3 that if an engineer assumes the limit given in RCS 112 then he will find that he will not go very far wrong.

The + and - signs in Fig. 6 should be omitted. With regard to the rule that film-type resistors should be used, this is generally correct. Fig. 8 was made up as a composite graph, based on the results of measurements on carbon composition, carbon composition film and cracked carbon film resistors. In all cases the r.f. performance of the film resistors was superior to that of the composition resistors.

If I may be allowed to comment on the last paragraph addressed to "Cathode Ray", it is useful to check by c.r.t. methods composition resistors used as noise standards with no current flowing, as sometimes additional noise may be superimposed from sources such as poor terminal contacts on slightly faulty resistors.

I would like to thank Dr. Bell for his valuable service in, as he says, "dotting i's and crossing t's" and I hope the above comments will also be found useful.

Great Malvern.

G. W. A. DUMMER.

## Fringe Area Reception

ON Sept. 16 an unsuccessful attempt was made to receive the Crystal Palace TV signals in America. At. the same time many thousand attempts to receive the same signal were being made in East Anglia and surrounding counties which were also unsuccessful. After six years of promises, it is disheartening to those who purchased receivers for some reception from Alexandra Palace to now have their receivers rendered nearly useless because of fading down to $10 \mu \mathrm{~V} /$ metre nearly every day. The south coast has received a better service at the expense of the northern limits of the area.

The B.B.C. charter makes some mention of bringing its services to the widest audience and the B.B.C. claims (I think) 98 per cent of the population served, which means one million without a TV service. Whilst not wishing to belittle the Corporation's magnificent efforts, a million is a lot of people and I suspect that signals of $100 \mu \mathrm{~V} /$ metre, are included in the service area. If so the B.B.C. is satisfied with very low picture standards and there must be few areas that average $100 \mu \mathrm{~V} /$ /metre without severe fading. In some areas of Huntingdon county a three element array at 50 ft will hardly guarantee one evening per week without loss of picture. Ponder on this those who have forgotten where the contrast control is situated.

A wide strip of land extending to the Wash has no satisfactory sound or television service of any kind, yêt the full licence fee is demanded by law from those who
valiantly try to obtain some entertainment. If fees were payable according to the service, would these forgotten (except on January first) areas be included in future plans?

St. Ives, Huntingdon. H.. S. KING.
IN some fringe areas, car ignition noise is beyond description in intensity.

On vision, it is comparatively easy, by attention to line sync-circuits and use of a simple diode limiter to eliminate 99 per cent of the effects, but on sound it is a different story. No amount of "limiting" seems to have much effect-I, personally, have installed three limiters, all at once!-yet experience shows that with f.m. noise reduction is fairly easy.

One is led to believe that the reason the U.S.A. is fairly free of complaints about ignition noise is due to the use of flywheel syne and f.m. sound, therefore I would like, on behalf of long-suffering fringe viewers in the south, to stake out a claim for, specifically, an f.m. sound carrier between Channels 1 and 2 .

Peacehaven.
R. G. YOUNG.

## Underwater Television

"DIALLIST'S" reference to underwater television (September issue) cannot pass without comment. Several instances where the medium was used occurred before 1951. Perhaps the first use was made in Germany prior to the last World War; certainly in 1947 it was used in connection with the atomic bomb explosion at Bikini Atoll. In 1948 the Scottish Marine Biological Association began preliminary work using an underwater camera for the study of marine life. Actually in 1949 a demonstration was given at the London Zoo using an E.M.I. C.P.S. Emitron camera.

The operation to locate the ill-fated Affray in 1951 was probably the next important use and also the first use as an aid to marine salvage. The Admiralty Research Laboratories at Teddington made the chamber to house the camera, which was a normal Marconi broadcast type about to be supplied to the B.B.C.

The first television camera to be designed purely for submarine use was the work of Pye, Ltd. The design was in an advanced stage just prior to the operation to locate the "Comet" off Elba. A suitable casing was designed for work down to 1,200 feet and the camera was subsequently used in this operation with great success.

Southampton. B. A. HORLOCK.

## SHORT-WAVE CONDITIONS Prediction for November



THE full curves given here indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during November.

Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.
.......... frequency below which communication should
BE POSSIBLE FOR $25 \%$ OF THE TOTAL TIME

-     -         - predicted average maximum usable frequency
———mequency below which communication shoulo
BE POSSIBLE ON ALL UNDISTURBED DAYS


# Pocket Transistor Receiver 

Long-wave Regenerative Circuit with Novel Stabilizing System
P. THARMA,* B.Sc (Eng.)

THE receiver described in this article is the result of an investigation to determine the smallest and the most economical design for a personal set. As the receiver was to be self-contained it was necessary to use a small ferrite rod aerial and this type of aerial, in turn, demanded a highly sensitive circuit. High sensitivity can be obtained either with a super heterodyne circuit, which requires about four transistors, or by using regenerative detection, when only two transistors are required. By suitable circuit design the regeneration can be made reasonably stable and gives good results.

The use of the OC71 transistor limits the frequency of reception to below about $400 \mathrm{kc} / \mathrm{s}$. Hencé the design of the receiver has been confined to reception of the Light Programme transmission on 200kc/s.

The useful range of this receiver with the ferrite rod aerial is about 120 miles from Droitwich, depending somewhat on local conditions. Adjustments become unduly critical for greater distances. The receiver is directional, having a figure-of-eight pick-up characteristic. An external capacitive aerial can be used where practicable and gives much greater sensitivity. Owing to variations of transistor characteristics with temperature, occasional readjustment of the controls will be necessary.
The receiver, which operates into a hearing-aid type earpiece, can be made not much bigger than a packet of 20 cig 3 rettes.

A complete, circuit diagram is given in Fig. 1. The first transistor functions as the regenerative detector, with signal input to the emitter, regeneration being provided by means of the feedback coil $L_{2}$. D.C. bias to the base is provided by means of the potential divider $\mathrm{R}_{8}$ and $\mathrm{R}_{3}$. Resistors $R_{1}, R_{2}, R_{3}$ and $R_{8}$ provide stabilization of the collector current, which is $100 \mu \mathrm{~A}$. Capacitor $\mathrm{C}_{7}$ decouples the base for r.f. and a.f. Capacitor $\mathrm{C}_{5}$ decouples $\mathrm{R}_{1}$ for r.f. only and provides stabjlization of regeneration, as explained later. Capacitor $\mathrm{C}_{6}$ decouples $\mathrm{R}_{2}$ for r.f. and a.f. The a.f. circuit design is conventional except for a small amount of positive current feedback which gives an increase of gain.

A Ferroxcube rod is used for the aerial. Grade A4 of this material has the highest useful permeability at $200 \mathrm{kc} / \mathrm{s}$ and is
*Mullard Ltd.
therefore specified. A fairly short rod has been used in this design to give ready portability. Much greater pick-up can, however, be obtained with a longer rod. An external capacitive aerial can be used by coupling directly to the tuned circuit via a $22-\mathrm{pF}$ capacitor. An earth should then also be connected.
A regenerative detector performs two functions: (a) detection, and (b) providing r.f. feedback to reduce the damping of the input circuit. Detection takes place in a transistor circuit at low currents due to non-linearity of the input characteristic and is most efficient at very low currents. However, the r.f. gain decreases with decreasing current. This decrease is gradual to $75 \mu \mathrm{~A}$ and rapid for lower currents. Satisfactory operation was obtained at a collector current of $100 \mu \mathrm{~A}$ using the transistor in the earthed-base mode.
Positive feedback to produce regeneration can be introduced in several ways. The sensitivity of the circuit increases as the feedback is increased towards the point of oscillation and the maximum usable sensitivity depends largely on the smoothness of controls and the stability of the circuit. Smooth regeneration control and adequate stability have been attained by careful coil design and the use of a stabilizing circuit, which may be of interest also in other applications.

In this circuit (see Fig. 1) the capacitor $\mathrm{C}_{5}$ is chosen so that only the r.f. is bypassed. This provides stabilization of regeneration.

Suppose that the input characteristic of the transsistor is similar to that shown in Fig. 2(a). This type of demodulator characteristic gives the same detection efficiency for all signal levels and the r.f. gain is constant provided the working point is at 0 .
fig. 1. Complete circuit diagram of the receiver.


The actual input characteristic of the transistor is approximately exponential, as shown in Fig. 2(b). This type of characteristic gives at all working points detection efficiencies and r.f. gains dependent on the signal amplitude.

Variation of detection efficiency results in a.f. distortion and is undesirable. A more important consequence when regeneration is used is the variation of r.f. gain with signal. If the circuit is adjusted near the point of oscillation for low signals, it will burst into oscillation for higher signals (due to rise in signal level or on peaks of modulation).

A resistance in series with the emitter makes the characteristic of Fig. 2(b) approach that of (a) but also reduces r.f. gain as well as detection efficiency. A large amount of regeneration is therefore required to get any sensitivity. This is both undesirable and difficult to obtain.
A resistance shunted by a capacitance ( $\mathrm{R}_{1}, \mathrm{C}_{5}$ in Fig. 1) where the capacitance is large anough to bypass only the r.f. frequencies, has a much smaller effect on the r.f. gain, thus making efficient regeneration possible. The a.f. voltage across $\mathrm{R}_{1}$ moves the operating point in phase opposition to the a.f. output provided by the demodulation process. This negative feedback greatly reduces the non-linearity of the latter process, and by so doing also reduces the variation in r.f. gain, which variation is proportional to the non-linearity of demodulation. This makes the characteristic of Fig. 2(b) approach that of (a) as far as dynamic operation is concerned. The system also tends to counteract variations in those characteristics of the transistor affecting gain. These advantages normally outweigh the loss in a.f. gain. However, occasional readjustment of controls may be necessary due to the influence of temperature.

The coils are wound on the Ferroxcube rod as shown in Fig. 3. This design takes into account $Q$ factor, coupling coefficients, optimum wire size and variation of transistor characteristics. One or other of the tappings gives better results and this can be determined by experiment.


Fig. 2. Transistor demodulation charocteristics: (o) an ideal curve giving linear demcdulation; (b) the octual characteristic with demodulation and gain non-lineor.


Fig. 3. Details of the coils wound on the Ferroxcube rod aerial.

## COIL WINDING DATA

All three coils are wound with 35 s.w.g. enamelled double-silk-covered wire on a core consisting of a Ferroxcube rod 4 in long and $\frac{1}{2}$ in in diameter (Grade A4, FXIO91).
$L_{1}=100$ turns wound multilayer over $\frac{1}{4} \mathrm{in}$.
$\mathrm{L}_{2}=57$ turns with tap at 40 turns, wound multilayer over $\frac{1}{4}$ in.
$L_{3}=17$ turns, single layer, close wound.
The coils are all wound in the same direction and are insulated from the core with Sellotape. Tuning coil characteristics: $Q=150, C=425 \mathrm{pF}$ at $200 \mathrm{kc} / \mathrm{s}$, unloaded.



Fig. 4. Constructional details of the receiver.
The following considerations guided the design. A value of $6: 1$ for the ratio of the number of turns in the tuning coil to the number of turns in the emitter coil gave optimum matching between the tuned circuit and emitter input. The reaction winding has to be tightly coupled to the emitter coil. Too large a number of turns on the reaction winding necessitates a small capacitance for $\mathrm{C}_{3}$ and makes the reaction control coarse and difficult to adjust and introduces a backlash effect on it. If the number of turns on the reaction winding is small it may not be possible to obtain regeneration close to the point of oscillation. From this it appears that the reaction winding has an optimum value for each transistor. Best results are obtained when the winding is such that the capacitance of $\mathrm{C}_{3}$ is about $300-400 \mathrm{pF}$ when adjusted for maximum sensitivity.

Transformer coupling is used between the detector and output stage. As the output impedance of the detector transistor is high, due to operation at low current, a ratio of $8.5: 1$ is used for the coupling transformer. The primary inductance must be high enough to give an adequate frequency response ( 38 H gives a drop of 3 dB at $300 \mathrm{c} / \mathrm{s}$ ).

An output transformer is used to match into the earpiece, which has an impedance of 250 ohms at $1,000 \mathrm{c} / \mathrm{s}$. The choice of the output transformer ratio is a compromise between gain and output power. A transformer ratio of $4.5: 1$ with a battery supply of 2.7 V gives good results, providing sufficient power for comfortable listening.

Positive current feedback is applied to the output stage via the resistor $\mathrm{R}_{7}$. If this is selected to give an increase of gain of about 10 dB , there is no serious effect on quality and the stability margin is adequate. The value of $\mathbf{R}_{7}$, required depends on the transistor V 2 and is of the order of $10 \Omega$. This positive feedback may be omitted if the receiver is intended for use in high field-strength areas. D.C. stabilization of the operating point of the output transistor is provided by means of the resistors $R_{4}, R_{5}$ and $R_{6}$.

In certain applications it may be preferable to feed the a.f. signal into an existing high-gain amplifier system, such as a hearing aid. Under these circumstances the output stage may be omitted and the a.f. signal taken directly from the detector stage. If a low impedance output ( $1,000 \Omega$ ) is required, this
could be taken from the secondary of the transformer $\mathrm{T}_{1}$. A high-impedance output can be taken from the junction of the transformer $\mathrm{T}_{1}$ and collector of V1 via a r.f. choke.
Regarding the battery, mercury type dry cells are best suited for use with this receiver. The good voltage stability of these cells allows the transistor to operate under optimum conditions during the whole battery life. The total drain is about 1 mA and cells such as the Mallory type TR152 would give about 350 hours life at a cost of less than 1/6th penny per hour.
Fig. 4 shows a possible layout of the components. The layout is not critical, except that the transformers $\mathrm{T}_{1}$ and $\mathrm{T}_{2}$ should be spaced far apart to prevent interaction. Trimmers $\mathrm{C}_{1}$ and $\mathrm{C}_{3}$ are preset, whilst trimmers $\mathrm{C}_{2}$ and $\mathrm{C}_{4}$ are adjusted in use. Small variable capacitors of the required sweep are not easy to obtain. Compression trimmers have been modified by replacing the existing trimming screw with a larger screw to which a knurled disc is attached. These are mounted so that the edges of the discs project through slots in the case.

For the setting-up procedure, the lower of the two tappings of the reaction winding should be tried first. The variable capacitors $\mathrm{C}_{2}$ and $\mathrm{C}_{4}$ are set in mid-position. The reaction capacitor $\mathrm{C}_{3}$ is set at maximum and the receiver switched on. If the circuit is in oscillation a whistle will be heard as the capacitor $C_{1}$ is varied, when the frequency of oscillation approaches the signal frequency. The reaction capacitor $\mathrm{C}_{3}$ is then gradually reduced and by alternate adjustment of $\mathrm{C}_{1}$ and $\mathrm{C}_{3}$ the signal can be tuned in. (If the circuit does not oscillate the higher tapping of the reaction winding should be tried.) The external aerial and earth are then disconnected. Some retuning and adjustment of the reaction control is then necessary to obtain optimum reception with the Ferroxcube rod aerial alone.
The author is indebted to L. H. Light for valuable advice in the design of the circuit.


ON-THE-SPOT fitting of car radio within the Greater London area is now provided by Rootes Ltd., who have equipped a van specially for the task. It is a combined installationdemonstration vehicle and is itself fitted with the new 20X Radiomobile receiver. Being on call (Ladbroke 3232) it obviates motor traders carrying stocks of radio parts.

The author's living room (reverberation time 0.46 sec ) showing placing of the loudspeakers for stereophonic sound reproduction.

AN APPRECIATION AND
SOME EXPERIMENTS IN PSYCHO-ACOUSTICS

By J. MOIR,* M.I.E.E.


## Stereophony in the Home

IN every key. cinema in the country a stereophonic sound film system is now installed, and as the sound quality that can be obtained from a proper stereophonic recording is so impressive it is natural to apply similar techniques to domestic sound reproducer systems. The recently released musical film "Oklahoma" is an excellent example of the results that can be obtained in the cinema from a stereophonic recording of music and singing.

The advantages of stereophonic sound reproduction are at least as great in the home as in a large hall, and the following notes are based on experience gained in running a two-channel stereo system in ordinary domestic surroundings. The views may be a little coloured by a professional interest in the design of stereo sound systems for cinemas.

It is not proposed to recapitulate the fundamentals of stereophonic sound reproduction, for this aspect has been adequately dealt with in previous contributions to Wireless World $\dagger$ but it will be remembered that two, three or five separate channels must exist between the recording studio and the reproducing room, the stereo performance improving as the number of channels is increased. Practical considerations restrict the number of channels to two when domestic equipment is considered.

Stereophonic recordings were made many years ago by Blumlein of E.M.I. using simultaneous lateral and vertical modulation of a single groove on a gramophone record, but magnetic tape has so many advantages over discs that all current British releases are on twin-track magnetic. tape. It seems likely that tape will become the standard storage medium though some of the smaller American companies issue a few discs having two recorded grooves.

In England the recordings available so far have been released on $\frac{1}{4}$ in magnetic tape by the E.M.I.

[^5]group under the "H.M.V." and "Columbia" labels and these are the only ones of which experience has been obtained. Without waiting for the end of the story it can be said at once that stereophonic reproduction in the home is a thrilling experience and one that resuits in a strong desire to "ask for more".

First a few words about the recorded tapes. The standard $\frac{1}{4}$ in tape provides space for only two tracks and a $30-\mathrm{mil}$ space (approximately) between them to minimize crosstalk. The dimensions are as in Fig. 1.

A stereophonic recording demands that the two recording heads be in an agreed position and that the replay heads be in the same relative position. This is essential if the recording system is to preserve the time differences that are so important to a good stereo performance. E.M.I. arrange their recording heads in line across the tape and it is thus essential that the replay equipment should also use heads with their replay gaps in line.

The microphone technique now employed in the E.M.I. is based on the Blumlein proposals, two directional-microphones being mounted one above the other and mutually at right angles. However, our present purpose is to discuss the results obtained under domestic conditions rather than the studio techniques.

## Equipment Used

Commercial equipment for playing "Stereosonic" tapes is available from several advertisers in Wireless World, but the price of a complete replay chain is a little on the high side for anybody on an engineer's salary; in consequence the writer's equipment shows signs of cinematic connections! For the enthusiast willing to assemble his own equipment several firms have available tape decks, including two compensated pre-amplifiers with an output in the region of one volt across 500 ohms. At a still lower price a tape deck can be obtained and the preamplifiers and main amplifiers assembled at home.

The tape reproducer used in my experiments is a

Ferrograph 2C/NF having in-line heads and an output of about one volt from the two internal preamps. The main amplifiers are some standard units used in the smaller cinemas and having frequency characteristics within 1 dB of each other over the whole audio-frequency range. Several types of loudspeakers have been tried, but it is still unfortunately true that the most expensive loudspeakers give the best reproduction. These particular examples are dual-channel, horn-through-the-centre-pole type (B.T.H. Type K10A) mounted in ported cabinets. They can be seen in the accompanying photograph of the living room. If the room is large enough highquality speakers of this type are ideal, but they are not essential, a point that will be referred to later. The seating, a settee and two easy chairs, was arranged across the room about $10-12$ feet away from the speakers.

With this arrangement the results obtained are outstandingly good and represent a much greater improvement over long-playing records than the 1.p. records show over the old 78 -r.p.m. recordings. The ordinary technical criteria, low harmonic distortion, wide frequency range, greater volume range, and low background noise, are inherently satisfied in a good magnetic tape recording and all the tapes tested set a standard much higher in these respects than any but laboratory examples of disc recording.

Surprisingly enough these characteristics received little comment from the 20-30 friends who have heard the results. What did receive comment from everybody, whether technically or non-technically minded, is the extraordinary improvement in clarity, the complete separation of orchestra and singers, the apparent size of the orchestra and the ease with which one can listen. This last point is a rather unsuspected and rather under-emphasized advantage of stereo recordings. One listening session ran into a couple of hours without the slightest sign of fatigue though the playing of a good l.p. record immediately brought the comment of "How hard that is to listen to after hearing the tapes ".

## Loudness Levels

The work of Somerville and Brownlees in England, and Chinn and Eisenberg in America has shown that the public have a marked preference for lower-thanoriginal sound loudness levels when reproducing music at home. This has been confirmed on many occasions in the present house, peak levels greater than about 85 phon always raising adverse comment from the other members of the family. After ten minutes playing of the first stereo tape it became obvious that peak loudness levels well above 85 phon were being experienced, so the sound level meter was brought out. This confirmed that the peak levels were in the region of 95-98 phon, a loudness level that would certainly arouse protests if the standard monaural system was being played. Even when disaster was courted by a direct question "What about the loudness?" only one listener in the group of six thought that "it was a bit loud".
A few words about some of the tapes that have been played "The Marriage of Figaro" by the Glyndebourne Opera Company (HMV SAT 1007) is an outstanding example of the advantages of a stereophonic recording. Normally I have no special liking for opera sung in Italian, but this tape was played at least ten times during one day for the sheer pleasure


Fig. 2. Alternative positions tried for the loudspeakers in a living room.
of hearing the performance with all the spatial characteristics of the original. The orchestra is a separate group spaced from the singers, and the principal singers appear against the choral background without the least confusion. This is a recording to demonstrate to friends who believe they have a high-quality disc recording system!

Orchestral music is almost equally impressive, the Classical Symphony (Prokofiev, No. 1 in D Major) SDT 1750, being most satisfying both technically and musically. This was another recording that was replayed at least ten times during one week-end, always bringing all domestic activity to a standsill. A long experience of high fidelity at home suggests that continued satisfaction is a rarity if records are being played.

It must not be thought that a stereo system is only suitable for "serious" music; Phil Green and his orchestra in "Interlude for Melody," Columbia BTD 701, is another outstanding example of the advantages of stereo recording, though my eleven-year-old daughter informs me that Phil Green and his band are a "square". I am not sure what this means, but I suspect that he plays music rather than merely making a row. Perhaps somebody more attuned to the younger mind or having access to a suitable dictionary can translate the expression into English. A fair number of other recordings have been played, all with real delight, but those quoted might make a good foundation for a tape library.

Some twenty people have listened to the equipment at various times, most of them being high-quality enthusiasts though diluted by a wifely accompaniment. Without exception all their comment has been extremely favourable, but perhaps the greatest compliment is that three ladies sat for more than an
hour listening to opera in Italian and music by Prokofiev and Sibelius without saying a single word to each other except in the intervals. In a very long experience of domestic high quality I have never known this happen when playing records. The explanation is probably fairly simple; the reproduction is so realistic that it would just be sheer bad manners to indulge in conversation.

Some of the more technical conclusions reached as a result of experiments are probably of interest. The loudspeakers used each have a fairly uniform polar distribution over an angle of about 60 degrees. Speakers having a less uniform angular distribution need to be placed more carefully but give very good results if this care is taken. The two ported cabinets were placed in the corners of the room, pointing down the length of the room and turned outwards to make an angle of about 30 degrees with the wall behind. The cone centres were then about 8 ft apart and the best listening position about 10-12ft away. At greater distances the stereo effects existed over the whole floor area out to the end of the room 21 ft away. Listening positions closer to the loudspeakers than the speaker spacing lose a good deal of the stereo impression.

Several types of speaker have been tried and in general the better the individual speakers, the better the stereo result. However, the quality of the speaker and particularly the frequency range covered appear much less significant when using a stereo system. Any enthusiasts thinking of trying domestic stereophony need not be deterred by the absence of two speakers of top-class quality. When listening to a real orchestra the quality is not judged by the frequency range of the instruments, and the same attitude develops when using a stereo reproducing system. The frequency range that is reproduced seems much less important than the sense of size and spatial distribution produced by the system.

## Loudspeaker Positions

Speaker positioning is a bit of a problem if the best listening position is to be integrated with normal domestic activity and the viewing of television. Speakers placed across the end of the room as in Fig 2 (a) give very satisfactory results, but are generally difficult from other points of view. A speaker placed at each side of the fireplace as in Fig. 2 (b) has advantages, in that it faces listeners seated round the hearth but the heating system needs to be a little better than "standard British" in order that the audience should be comfortable eight or ten feet from the fire. Seating positions outside the area demarcated by the speaker positions are not very satisfactory. The room used-for the majority of the tests is fairly good acoustically (measured reverberation time 0.46 seconds at $500 \mathrm{c} / \mathrm{s}$ ) but the acoustics of the listening room appear very much less important with a stereo reproducer system than with a monaural system. The acoustics of the recording studio are the predominating factor, a point well brought out when listening to the Glyndebourne recordings. Room size also appears to be a less significant factor, very good results being secured in a room only $12 \mathrm{ft} \times 9 \mathrm{ft}$ with the speakers across the end of the 9ft wall.

One very important factor is the relative volume level from the two speakers. Any 2-channel stereo system must depend for its main effects on the
intensity difference that is produced at the two ears, the intended result being secured by the correct positioning of the microphones at the recording studio. A reproducer system having separate gain controls in each channel allows the stereo result to be completely spoilt by incorrect setting of the gain controls. The ideal arrangement is undoubtably to have a single knob with the two channel controls ganged on the same spindle, though each channel must have a separate gain control that can be pre-set to equalize the acoustic output from the two channels. From this point of view it would be a great convenience to have available a " balance tape" with short bursts of $1,000 \mathrm{c} / \mathrm{s}$ tone recorded alternately on each track. This would allow the acoustic outputs of the two speakers to be balanced and the gain control settings noted for future use. Such "loudspeaker balance" films are available to the cinema engineer. (A loop of tape only 3 or 4 ft long is quite adequate.) The necessity of balancing, in some manner, cannot be too strongly emphasized, for lack of balance leaves the orchestra and all the singers on one side of the room, completely ruining any stereo effects.

The amplifiers used should have similar characteristics, a tolerance of $\pm 1 \mathrm{~dB}$ over the frequency range being suggested. If variable tone controls are fitted these need to be set to give the same overall frequency characteristic, or the orchestra tends to be concentrated towards the loudspeaker having the best highfrequency response. Differences in frequency characteristic of more that 2 or 3 dB lead to a peculiar "stretching" of the orchestra, while the individual instruments move about the stage as the player moves up and down the musical scale.

Finally about the future. In the cinema world stereophonic sound reproduction has been seriously jeopardized by the widespread release of films with identical sounds on all three tracks though advertised as having "stereophonic sound". In many films these have been dubbed on to the magnetic tracks from a single original photographic track. This is killing the goose with a vengeance. It is hoped that no one will attempt to foist such counterfeit on the domestic market.

In my opinion, stereophonic tape recordings of the present standard are sufficiently outstanding to ensure that they will be the accepted practice for all domestic high-quality systems in perhaps a couple of years time. Cost remains an obstacle at present, but it is to be hoped that prices will fall steeply as the demand rises.

## APPRENTICESHEPS

AN outline of the opportunities open to graduates and others in the Philips organization in this country is given in "Careers in Philips." It covers mainly the opportunities in the manufacturing organization but also touches upon the openings for technical staff in the commercial departments.

Two brochures available from Metropolitan-Vickers cover the apprenticeships which the company offers to graduates ("The Training of the Professional Engineer") and public, secondary grammar and technical schoolboys ("From School to Professional Engineering ").
"Opportunities in Electronics for University Graduates" is the title of a booklet issued by Mullard. It outlines the opportunities open to science graduates in the company's research laboratories and production organizations.

# Transistor R.F. Amplifiers 

2-PRACTICAL CIRCUITS WITH NEUTRALIZATION AND AUTOMATIC GAIN CONTROL

By D. D. JONES,* M.Sc., D.I.C.

(Concluded from page 496 of October issue)

IT was mentioned in Part I, last month, that if a voltage appears at the output terminals of a transistor amplifier, internal feedback due to $r_{b o}$ and $\mathrm{C}_{c}$ results in a voltage also appearing at the input terminals. Consider a common-base amplifier working into a load $\mathbf{Z}_{\mathrm{L}}$. If an a.c. current I is made to flow in the input circuit a voltage approximately equal to $\alpha \mathrm{IZ}_{\mathrm{L}}$ will appear across $\mathrm{Z}_{\mathrm{L}}$; the phase relationship between this and the applied signal will depend on the phase change introduced by the complex nature of $\alpha$. Part of this voltage will now be fed back into the input circuit via $\mathrm{C}_{c}$ and $r_{b o}$ and will modify the input impedance of the amplifier; the extent of this modification depends on the


Fig. 7. Method similar to that shown in Fig. 6, but with $r$. and $C$, taken into account.
magnitude of $\mathrm{Z}_{\mathrm{L}}$ as well as on $\alpha, \mathrm{C}_{\mathrm{c}}$ and $r_{b 0}$. The input impedance does therefore depend on the load; as was mentioned in the introduction this effect does not arise in thermionic valves until the frequency is comparatively high.

If a tuned circuit forms part of the load, $\mathrm{Z}_{\mathrm{L}}$ will vary with frequency; this can give rise to considerable variations in input impedance over the bandwidth of the tuned circuit (and hence, of the amplifier). It is desirable to introduce a subsidiary network into the amplifier circuit to produce another feedback component which cancels out that produced by the transistor.

Neutralizing networks of this kind play an important part in present-day transistor circuits.

There are a number of possible neutralizing arrangements, but so far only two or three of them have been extensively used. A useful method both for seeing the necessity for neutralization, and also for aligning the circuit itself, is to measure the input impedance of an amplifier as a function of frequency over the pass band; the amplifier load includes the tuned circuit in these measurements.

In the common-base amplifier the input impedance can increase to very high values in the pass band; it can in fact become infinite, change its sign and give rise to oscillations. In the case of the common-emitter amplifier, the input impedance decreases over the pass band and can again become negative and cause oscillations. In properly neutralized amplifiers there should be no significant changes of input impedance over the pass band.

The circuit shown in Fig. 5 is found to be satisfactory for neutralizing the common-base amplifier. Any voltage appearing across the output terminals will give rise to a voltage across the input terminals due to internal feedback provided by $r_{b o}$ and $\mathrm{C}_{c}$; however, a voltage will also be fed back into the input terminals via $\mathrm{C}_{n}$ and $r_{n}$ and if

$$
r_{n} \mathrm{C}_{n}=r_{b o} \mathrm{C}_{c}
$$

these fed back voltages will be equal and will cancel each other out.
The circuit shown in Fig. 6 can be used for the common-emitter amplifier. In this case the capacitance $\mathrm{C}_{\mathrm{c}}$ feeds a voltage back into the input circuit; by means of the phase inverting transformer and the capacitor $\mathrm{C}_{f}$, an out-of-phase voltage of equal amplitude is also applied to the input circuit. The effectiveness of this method depends on $r_{b o}$ being low. A more satisfactory result can be achieved if the circuit of Fig. 7 is used; however, since $r_{e}$ varies rapidly with emitter bias current it is found that this bias must be kept reasonably constant for the method to be effective.

Any practical amplifier is expected to give satis-
factory performance over a range of operating conditions. A fair test of the performance of any neutralizing circuit, therefore, is to observe its variation with operating conditions, e.g., with change of supply voltage.
When using GET4 transistors in $465-\mathrm{kc} / \mathrm{s}$ i.f. amplifiers, it is found that the common-base circuit with the neutralizing method found to be adequate. neutralizing network become. is employed. a turns ratio of about $20: 1$ is used, though this critical neutralization. of Fig. $9(\mathrm{~b})$ is often useful.
shown in Fig. 5 can withstand large changes in supply voltage. It is also found that the amplifier is more stable if $r_{n} \mathrm{C}_{n}$ is slightly less than $r_{b o} \mathrm{C}_{c}$; using the mean values of $r_{b o}$ and $\mathrm{C}_{c}$, obtained from the production "spread" of the transistor, is also

The higher the dynamic impedance of the tuned circuit used, the more critical does the design of any

Fig. 8 shows a two-stage $465-\mathrm{kc} / \mathrm{s}$ i.f. amplifier using GET4 transistors. A dynamic impedance of about $50 \mathrm{k} \Omega$ is used in this case to avoid the necessity for very accurate neutralization. A gain of about 36 dB is obtained with such an amplifier. This gain is considerably lower than would be obtained if transistors having much higher $\mathrm{M}_{a}$ were used; however, it does show that using transistors in what are undoubtedly "marginal" circuits, useful gain can be obtained if neutralization

In order to obtain as much gain as possible it is necessary to use interstage transforming circuits. The transformation ratio depends on the dynamic impedance of the tuned circuit forming the collector load of one transistor and the input. impedance of the next stage. In the amplifier shown in Fig. 8 could be increased at the expense of rather more

Of the various types of matching circuits that can be used, that shown in Fig. 9(a) is widely employed in practice. In experimental work the circuit


Fig. 8. A two-stoge $465-\mathrm{kc} / \mathrm{s}$ i.f. omplifier using GET4 tronsistors.


Fig. 9. Interstage coupling networks.
Since $\mathrm{C}_{c}$ and $\mathrm{C}_{n}$ form part of the tuned circuit it is important to remember that $\mathrm{C}_{\mathrm{c}}$ varies with supply voltage. To minimize this effect, the tuning capacitor should be large.
It is often desirable that the gain of an amplifier can be automatically adjusted according to the mean level of the input signal. Unless the gain is controlled in this way the amplifier may be overloaded at high levels of input signal.

Automatic gain control can be achieved in two ways. In the first method, the actual d.c. voltage at the collector is reduced as the mean signal level increases; both the current gain factor and the collector capacitance vary with collector voltage and contribute to a reduction in the gain of a tuned amplifier. The second method is based on reducing the emitter bias current as the mean signal level


Fig. 10. Automatic gain control of o common-emitter amplifier.
increases; this technique is used more extensively than the former and is therefore discussed in greater detail.

In Part I of the article it was mentioned that the input impedances of the common-base and commonemitter amplifiers, when working into low load impedances, are given by

$$
\begin{aligned}
& \mathbf{Z}_{i n}=r_{e}+r_{b o}(1-\alpha) \text { for common-base } \\
& \mathbf{Z}_{i n}=r_{b 0}+\frac{r_{e}}{1-\alpha} \text { for common-emitter. }
\end{aligned}
$$

( $\mathrm{C}_{e}$ being ignored for the moment). When the load impedance is comparatively high it is found that the parameter $r_{e}$ still plays a dominant role in the expression for input impedance. It was also mentioned earlier that $r_{e}$ is inversely proportional to the emitter bias current $I_{e}$. If this current is varied $Z_{i n}$ will be varied in both amplifiers.

The emitter bias current, $\mathrm{I}_{e}$, is related to the base bias current, $I_{b}$, by the expression

$$
I_{B} \sim \frac{1}{1-\alpha} I_{b}
$$

Since $1 / 1-\alpha$ is, typically, 40 it will be seen that large changes in $\mathrm{I}_{e}$ and hence in $r_{e}$ can be achieved
by small changes in $\mathrm{I}_{\text {b }}$. In both common-base and common-emitter amplifiers it is general to arrange the d.c. bias circuit so that the value of $I_{b}$ can be varied.

Fig. 10 shows a circuit used experimentally for providing a.g.c. In this, the output from the amplifier is rectified by diode $D_{1}$ to give a positive output voltage. This is then fed back to the base of the first amplifier and reduces the base bias current (which is negative in sign). The emitter resistor $R_{k 1}$ should be as low as possible because it produces an undesirable increase in the d.c. input resistance of the amplifier; the a.g.c. may be regarded as being substantially d.c. The resistance of $R_{D}$ is made fairly large ( $\sim 20 \mathrm{k} \Omega$ ) and, as a result, much of the positive bias voltage $\mathrm{V}_{\mathrm{EE}}$ is developed across it. This biases the diode in the reverse direction, thereby giving rise to a delayed a.g.c. characteristic. As the signal input to the diode is increased, the positive voltage between emitter and base, and hence the gain of the amplifier, is reduced.

In amplifiers where a transistor is used as a second detector a similar principle can be employed to obtain a.g.c.

## BOOKS RECEIVED

Radio Research 1955. Report of the year's work by the Radio Research Board (D.S.I.R.), which included phase changes in ground waves of low frequency, propagation at v.h.f. and u.h.f. (including direction finding at these frequencies), measurement of atmospheric noise and investigation of ferrite and semi-conductor materials. Pp. 56; Figs. 8. Price 3s 6d. Her Majesty's Stationery Óffice, York House, Kingsway, London, W.C. 2 .

Automatic Integrator for Determinating the Mean Spherical Response of Loudspeakers and Microphones by A. Gee, M.A., and D. E. L. Shorter, B.Sc.(Eng.), A.M.I.E.E. No. 8 in the series of Engineering Monographs published by the B.B.C. describes the use of a modified kilowatt-hour meter for integrating the response while plotting the polar diagram on an automatic recorder. Pp. 16; Figs. 10. Price 5s. B.B.C. Publications, 35, Marylebone High Street, London, W.1.

Frequency Modulation Engineering by Christopher E. Tibbs, M.I.E.E., M.Brit.I.R.E., and G. G. Johnstone, B.Sc. Revised second edition of a comprehensive survey of the principles and practice of transmitting and receiving f.m. signals. Includes mathematical treatment to the level required for the design of components and systems. Pp. 435; Figs. 254. Price 45s. Chapman and Hall, 37, Essex Street, London, W.C.2.
Radio (Vol. 3) by J. D. Tucker and D. F. Wilkinson, B.Sc.(Eng.), A.M.I.E.E. Text book covering the syllabus of the Radio III examination of the City and Guilds of London Institute. Pp. 249; Figs. 237. Price 12s 6d. English Universities Press, Ltd., 102, Newgate Street, London, E.C.l.

Radio Servicing (Vol. I. Electrotechnology) by G. N. Patchett, B.Sc., Ph.D., M.I.E.E., M.Brit.I.R.E. Fundamental electrical concepts underlying the theory and practice of radio (including television) servicing. Designed to cover the syllabus of the City and Guilds and R.T.E.B. examinations. Pp. 80; Figs. 84. Price 5s. Norman Price (Publishers), Ltd., 283, City Road, London, E.C.1.

Hi Fi Loudspeakers and Enclosures by A. B. Cohen. Lucidly illustrated analysis of the construction and
principles of operation of loudspeakers, cabinets and folded horns (principally of American design and manufacture) with an appendix giving constructional details of 18 representative types. Pp. 360 ; Figs. 183. Price 37s 6d. Chapman and Hall, 37, Essex Street, London, W.C. 2 .

Introduction to Colour TV M. Kaufman and H. E. Thomas. Second edition of a general description of the NTSC system with stage-by-stage analysis of a receiver and a full circuit diagram. Pp. 156; Figs. 81. Price 25 s . Chapman and Hall, 37, Essex Street, London, W.C. 2 .

Mandl's Television Servicing by M. Mandl. Revised second edition of a teaching manual for students with new material on transistors and printed circuits. Illustrated by examples from American practice. Pp. 460 ; Figs. 314. Price 45 s 6d. The Macmillan Company, 10, South Audley Street, London, W.1.

An Approach to Modern Physics by E. N. da. C. Andrade, D.Sc., Ph.D., LL.D., F.R.S. Non-mathematical survey for the layman of recent developments in theoretical and experimental physics. Pp. 232; Figs. 7 and 16 plates. Price 25s. G. Bell and Sons, Ltd., York House, Portugal Street, London, W.C.2.

From Microphone to Ear by G. Slot. Philips Technical-Library Popular Series introduction to modern sound recording and reproducing technique. Deals primarily with disc recording but includes a chapter on magnetic tape. Pp. 169; Figs. 118. Price 17 s 6 d . Cleaver Hume Press, Ltd., 31, Wrights Lane, London, W. 8.

Les Antennes by L. Thourel. Comprehensive mathematical treatise on aerials with particular emphasis on decimetre and centimetre wavelengths. Includes chapters on horns, slots, and lenses. Pp. 440; Figs. 252. Price 4,800 Fr. Dunod, 92 rue Bonaparte, Paris, 6.

Informationstheorie Vol. 3 of "Nachrichtentechnische Fachberichte." A collection of 12 papers by authors of many nationalities on applications of information theory, with summaries in English and German. Pp. 118; Figs. 125. Price 16.50 DM. Verlag Friedr. Vieweg and Sohn, Postfach 185, Braunschweig, Germany.

The Second International Congress on $A$-ouss:ics was held at Cambridge, Massa=husetts, in June this year in conjunction wlih the Annual Meeting of the Acous isal Society of America. These Congresses, which are sponsored by the Interna ional Commission on Acous ics, one o the specialized commissions of the In erna lonal Union of Pure and Applied Physics under UNESCO, are being held at intervals of three years. The first was held a: Delft in 1953.

# International Acoustics Conference 

Some Impressions of this Year's Meeting in the U.S.A.

FROM A CORRESPONDENT

THE opening ceremony and the first two days' rechnical sessions were held at the Massachusetts Institute of Technology. The following two days' technical sessions took place in Harvard University and the fifth and final days' meetings were again at M.I.T. In addition to the technical sessions, some visits, a concert and a banquet were arranged for the delegates.
The meetings at Harvard were of particular interest because of the associations with W. C. Sabine who originally postulated the concept of reverberation time. Some of the meetings were held in the Sanders Theatre which was one of the auditoria in which he conducted his experiments. His original reverberation chamber was on view, as were some of the famous cushions which he used in providing absorption experimentally. The concert was given by the Boston Symphony Orchestra in the Boston Symphony Hall which is an important acoustic design by W. C. Sabine.
Because this Congress was held in conjunction with the Annual Meeting of the Acoustical Society of America the attendance was much greater than at Delft. In all there were about 900 delegates and almost 300 papers in various classifications. In addition to the delegates there were, of course, associates, bringing the total to the region of 1,200 . The variety of papers was great, covering all the aspects of acoustics. Typical classifications were as follows:-

> Architectural Acoustics
> Musical Acoustics
> Sneech Analysis and Synthesis
> Physical Acoustics and Sonics
> Loudspeakers and Sound Reproduction
> Noise Control and Measurements
> Bioacoustics Acoustics
> Geophysical Ach
> Radiation and Scattering of Sound

On the evening of the first day a demonstration of speech analysis and synthesis was given by W. Lawrence of S.R.D.E., which was exactly similar to the demonstration he has already given in this country to the I.E.E. and to the Acoustics Group of the Physical Society. Following this another demonstration of speech synthesis was given by G. Fant of Stockholm which, although in some respects it did not reach the standard of the S.R.D.E. demonstration, appeared to indicate a more flexible system
which might be capable of great development. A very humorous finale to these demonstrations took the form of a duet sung by both pieces of equipment.

For the benefit of all the delegates, a survey paper was read by Georg von Békésy on the mechanics of the cochlea. Békésy, who has now worked at Harvard University for many years, is well known for his investigations into the theory of hearing.

In the field of loudspeakers and sound reproduction various interesting papers were read. For example, papers were given on "Directional Loudspeakers for Sound Reinforcement" by S. Hill (S.T.C.); "The E.M.I. Stereo Recording and Reproducing Systems " by G. F. Dutton; " Displacement Pickup for Measuring the Motion of a Loudspeaker Cone at Many Points" by K. R. McLachlan. A paper was also read on the latest apparatus for the alteration of the playback time of a sound record by A. M. Springer, who now claims that this equipment, which is basically similar to the "Tonschreiber B" used by the Germans during the war, can expand the time of reproduction of speech and music by as much as 200 per cent and give satisfactory compression of 50 per cent. An interesting paper was read by R. Kirk on "Learning, A Major Factor Influencing Preferences for High-Fidelity Reproducing Systems ". This paper describes an attempt to evaluate the effect of experience on the opinions of subjects listening to high-quality electro-acoustic reproducing systems. Two groups were used; one listening to music reproduced between 30 and $15,000 \mathrm{c} / \mathrm{s}$, while the other group listened to 180 $3,000 \mathrm{c} / \mathrm{s}$. After they had listened for two hours a week for 13 weeks, tests were carried out which seemed to show that the subjects preferred the particular bandwidth to which they had been listening.

Papers on architectural acoustics reflected the work going on at present to provide effective objective means for measuring the properties of enclosures, but developments are so slow that it will be many years before complete reliance on objective methods will enable satisfactory designs to be carried out.
In the field of musical acoustics several papers described methods of synthesizing sounds, which in some cases were intended to represent normal musical instruments and in other cases to produce sounds entirely different. Some workers are con-
centrating on the analysis of musical sounds mainly with the object of being able to synthesize them.

A considerable number of papers were devoted to the study of problems of noise and its measurement. This aspect of acoustics assumes greater importance every day because of the new sources of noise, such as jet aircraft, which are now becoming so disturbing that they cannot be ignored. In the case of buildings, the desire to economize leads to lighter forms of construction which are sometimes quite inadequate for sound insulation.

This Congress was very successful in bringing
together all the workers in the various fields of acoustics. The delegates enjoyed the proceedings and appreciated the excellent organization. It is perhaps unfortunate that the large number of papers to be read reduced the time for discussion, a point which requires consideration in planning future congresses. Most of the papers read are to be published either in Acustica or in the fournal of the Acoustical Society of America.

At the concluding session it was announced that the Third International Congress on Acoustics will to be held at Stuttgart in 1959.

# Amateur Television 

Progress in Colour Transmission

SINCE the last general report on amateur television published in Wireless World over two years ago, steady progress has been maintained, resulting in some spectacular results. Perhaps the most interesting of these is the work on colour television. As already reported, C. G. Dixon, of Ross-on-Wye, obtained first results over a closed circuit with a home-made frame-sequential colour camera in early 1954, using 150 -line 100 -frames per second standards. Further development of the colour gear produced a simple colour slide scanner and colour bar generator of a reasonably transportable size. All this equipment was shown in use (closed circuit) at the 1955 British Amateur Television Club Convention, and as a result it was arranged that as soon as practicable, a colour transmission over the air should be tried. By early spring of this year, reliable two-way television transmission had been established between G2WJ/T at Dunmow and G3CVO/T at Chelmsford. Mr. Dixon's colour bar generator and rotating disc monitor were later incorporated in the monochrome apparatus at these two stations, and on April 7th successful colour patterns were transmitted between them. Mr. Dixon was unable to bring the colour camera for this test, but plans are in hand to use it for the transmission of colour pictures from Ross-on-Wyy to Birmingham once the path is established.

Naturally, much interest has been aroused by these tests, and although the system is not compatible, it is relatively easy for the amateur to use. Simultaneous colour systems, such as the " British-N.T.S.C.," are, so far as amateurs are concerned, at present limited to slide and film scanning, or to flying-spot studio scanning. On the other hand, simple framesequential receivers built to receive N.T.S.C.-type transmissions can easily be modified to receive amateur colour transmissions. For this reason, several colour monitors are being built, especially in the London area.

On the r.f. side, very great strides have been made. Following the continued success of the G2WJ/T (Dunmow) to G3GDR/T (Abbots Langley) trans-

By M. BARLOW (G 3CVO)


Television links esţablished by amateurs in East Anglia.
missions, admittedly over a very good v.h.f. path, contact has been established between six stations in East Anglia. Due to siting difficulties, not all of these stations can work all the others, and attempts are being made to relay pictures from one station to another. Unfortunately, the size of aerial required on some of the more difficult "hops" makes it awkward to mount two arrays at a reasonable height, with consequent reduction of signal-to-noise ratio. Similarly, attempts to work simultaneous duplex between two stations-a realization of a "television-telephone"-have not so far succeeded. It should be explained that all the vision signals are trans-
mitted in the $430-440 \mathrm{Mc} / \mathrm{s}$ region, with the sound channels, normally duplex, on lower frequency amateur bands such as $1.9 \mathrm{Mc} / \mathrm{s}$ or $145 \mathrm{Mc} / \mathrm{s}$. Attempts to space the sound channel the correct $3.5 \mathrm{Mc} / \mathrm{s}$ below the vision carrier are only successful over the better paths, as otherwise the loss of vision signal-to-noise ratio is too great to be tolerated as the receiver bandwidth is increased. To overcome this, and for weak-signal reception generally, it is becoming common to lock the amateur TV equipment to B.B.C. or I.T.A. synchronizing pulses at each end of the path; considerably weaker signals can then be resolved quite easily.
There are now 23 amateur television stations licensed, and 14 of these have actually radiated pictures. In Birmingham, G3KBA/T is in a very bad location, surrounded by tall steel-framed buildings, and in order to send pictures from his 16 mm telecine scanner to $\mathrm{G} 3 \mathrm{KQJ} / \mathrm{T}$ at Wolverhampton, a relay via a nearby station is necessary. G3KFH/T at Worthing has put out some test transmissions, and hopes to send pictures across the channel. With G3BLV/T (Sunderland) and G2DUS/T (Baldock) thinking in terms of portable television transmitters, no doubt the present record of 38 miles will soon be broken.

As was to be expected, the photo-conductive pickup tube has replaced all other camera tubes as the first choice for the amateur, on account of its sensitivity, small size and modest power and scanning requirements. Cameras are built along similar lines to industrial TV cameras, often with lens turrets and built-in viewfinders for which second-hand projection tubes run at reduced e.h.t. are popular. Existing cameras of the image iconoscope and image orthicon types are still used, of course, but their size is a distinct disadvantage in the average small workshop. Some surplus American airborne TV cameras have also been used with success, but the tube sensitivity is very low. With photo-conductive tubes, normal room lighting is usually sufficient to produce excellent pictures.
Flying-spot scanning remains the standby for most amateurs; newer blue-trace c.r. tubes enable firstclass pictures to be produced for a very small outlay. Tubes in domestic television sets can be pressed into service to give fair results, and considerable attention has been given to the problem of scanning cine film. Thus one 16 mm scanner consists of an unmodified Kodascope projector, with a $9-$ in MW22-14 tube as the scanner, using a 4in by 3in raster. The projector is driven by a synchronous motor, but asynchronous running is possible with slight flickering. A 931A photocell is used, with gamma correction in the video amplifier to correct for receiver tube characteristics. More favoured are the types of scanner in which the film is wrapped round a glass or perspex polygon in the same way as in a simple film editor. This system has the very great advantage for the amateur that a picture is produced at all film speeds without any synchronizing troubles. Scan reversing switches are normally fitted for quick saving of face when the film has been incorrectly loaded! For those with photo-conductive-tube cameras, the problem is much simpler; with the camera pointed at the projector (suitably dimmed), very reasonable pictures are produced with no synchronization whatsoever.

Miniaturization and economy in components are pressed to the limit, and several members are using

BRITISH AMATEUR TELEVISION STATIONS

| Call sign | Location | Vision |  | Sound Frequency (Mc/s) |
| :---: | :---: | :---: | :---: | :---: |
|  |  | Frequency (Mc/s) | e.r.p. |  |
| G2WJ | Dunmow, Essex ... | 436.0 | 250W | 432.5 or 145.3 |
| G3GDR | Abbots Langley, Herts. | 434.1 | 250W | 434.1 or 144.7 |
| G2DUS | Baldock, Herts. ... | 434.5 | 250W | 434.5 or I 44.5 |
| G3KKD | Ely, Cambs. ... | 434.1 | 2W | $434.1$ |
| G3KOK/T | Bishops Stortford, Herts | 437.0 | 60W | 433.5 |
| G3CVO | Chelmsford, Essex ... | 430.36 | 200W | 426.86 or 1.97 |
| G3KBA/T | Birmingham, Warwick. | 436.8 | 50W | 426.86 or 1. 97 |
| G3BLV | Sunderland, Durham | 435.1 |  |  |
| G5ZT | Plymouth, Devon ... | 427.0 | 50W |  |
| G3FWF | Belfast, N. Ireland | 437.75 |  | 434.25 |
| G3CTS | Norwood, London ... | 427.0 | 1 kW | 423.5 |
| G3KRA/T | Chelmsford, Essex .... | 442.0 | 2W | 438.5 |
| G3KQJ/T | Wolverhampton, Staffs. | 438.75 |  | 435.25 |
| G3KFE | Enfield, Middx. ... |  |  |  |
| G3JVO/T | St. Albans, Herts. ... | 445.5 | IOW |  |
| G3BAY | Leicester ... |  |  |  |
| G3FNL G3KPX/T | Upminster, Essex ... | 445.0 | 250'N | 441.5 |
| G3KPX/T G3AST | Maidenhead, Berks. ... Luton, Beds. ... ... | 434.25 | 50W | 144.75 |
| G3KFH/T | Worthing, Sussex ... | 433.0 | 50W |  |
| G3ACK | Blyth, Northumb. ... | 432.6 |  |  |
| G3LCM/T | Coulsdon, Surrey ... | 434.7 | 25 W |  |

Amateurs with a general transmitting licence add the suffix " $/ T$ " when transmitting television.
a simple frame pulse of one half-line duration rather than the full B.B.C.-type waveform. For the latter, some 14 valves are usually required, and the possibilities of transistors are beginning to be explored. As an idea of the size of a typical amateur TV equipment, it may be mentioned that G2DUS/T (Baldock, Herts) has in a rack 36 in high by 15 in wide the following units: 405 -line sync generator, $45-\mathrm{Mc} / \mathrm{s}$ oscillator and distribution amplifier, vision mixer, Staticon camera control unit, monoscope unit (Test Card C) and pattern generator, and power supplies. The camera is 10 in long by 4 in wide and 6 in high, and is fitted with $75 \mathrm{~mm}, 35 \mathrm{~mm}$ and 25 mm lenses. With the addition of a domestic TV receiver, a complete closed circuit television station is available.

With a view to informing and instructing amateurs in the problems of television transmission, booklets, a film and six instructional tape lectures have been produced. In addition, B.A.T.C. groups have been formed in Southampton, Worthing, North and South London, High Wycombe, Romford, Chelmsford and Birmingham, where meetings are frequently held. There is little doubt that we are in this country-well ahead of everyone else in this particular field, although there are active enthusiasts in the U.S.A., the Netherlands, South Africa, Australia and New Zealand.

Future developments are difficult to prophesy; a link between London and Birmingham, and between the United Kingdom and the Continent, are obvious targets. In the meantime experiments continue with a microwave link, pulsing banks of fluorescent coloured tubes for frame-sequential work in colour, simpler sync generators, better film scanners and more effective transmitters and receivers. No doukt when G3CTS /T, the station of the Television Society, at Norwood, starts regular transmissions this autumn, more and more people will be tempted to try it for themselves.

A representative selection of the equipment mentioned in this article will be shown in operation at the third Amateur Television Convention to be held on October 27th at the Bonnington Hotel, Southampton Row, London.


Fig. 1. The effect of insufficient tape speed control is shown at ( $a$ ) both on the received picture itself (a small section) and on the horizontal test bars to the right. At (b) is shown the improvement effected when the control equipment is in operation

## RADIO PICTURES ON TAPE

DURING the Olympic Games this winter, some of the news pictures arriving from Australia by radio phototelegraphy are likely to have undergone the intermediate process of storage on magnetic tape. The storage equipment will be operating at Singapore, which is the normal Cable and Wireless relay point on the "eastabout" route between Australia and the U.K. The purpose of the scheme is to facilitate the transmission of pictures when radio conditions are not good enough on both sections of the route simultaneously for direct relaying. Quite often when conditions are good on the first section they are bad on the second, and vice versa. With the intermediate storage, however, it will be possible to send a picture from Australia to Singapore even when radio propagation is bad between Singapore and Britain-then to forward it over this second section when the conditions improve.
In the past any storage has been done simply by printing the picture at the relay station and re-scanning it when required. This, however, tended to degrade the picture quality, mainly because of interaction between the two scanning patterns, and Cable and Wireless considered it desirable that the signal itself should be stored, preferably on magnetic tape. Here, the main difficulty arises from possible variations in the speed of the tape. Any such variations between successive scan lines* of the picture produce timing errors, which are particularly noticeable from the ragged reproduction of horizontal edges on the received: picture, as shown in Fig. 1(a).
To overcome this problem the Cable and Wireless tape storage equipment incorporates an automatic system of speed control comparable with an a.f.c. loop. This makes use of a stable reference frequency of $2 \mathrm{kc} / \mathrm{s}$ from a crystal oscillator, which is recorded on one track of the tape while the other track is taking the incoming picture signal. (An orthodox twin-track tape deck is used, running at 13 inches per second.) On playback this reference frequency
*Which are, of course, vertical on the drum scanning system used and are, in fact, 100 lines to the inch.
is compared with the original from the oscillator, and any difference resulting from a change in speed is used to produce a correction signal which controls the tape drive until the frequency difference is reduced to zero. Since the picture information on the tape and the reference frequency bear a constant relationship to each other (being recorded simultaneously) any correction applied to one means a correction to the other.
The actual comparison of frequencies is done by an electromechanical system. To begin with, the normal tape-drive motor is replaced by a phonic motor which, for the recording operation, is energized from the $2-\mathrm{kc} / \mathrm{s}$ crystal oscillator through a power amplifier. On playback, the recorded reference frequency from the tape and the output from the crystal oscillator are both amplified and used to operate two further phonic motors which drive the two input shafts of a differential gear. The cage, or output shaft, of the differential then rotates with a speed and direction corresponding to the difference between the two input speeds (and hence frequencies).
This "error signal" rotation, amounting to perhaps $1^{\circ}$ of angular movement, is used to drive a variable capacitor which controls the frequency of one of a pair of $100-\mathrm{kc} / \mathrm{s}$ oscillators in a b.f.o. arrangement. The beat frequency obtained is arranged to be $2 \mathrm{kc} / \mathrm{s}$, and this, after amplification, is used to energize the tape-drive phonic motor. Any error signals from the differential gear will cause this $2-\mathrm{kc} / \mathrm{s}$ beat to be varied (a matter of 3 or $4 \mathrm{c} / \mathrm{s}$ per degree of angular movement) and as a result corrections are applied to the speed of the tape-drive phonic motor in the right sense to reduce the error signals to zero.
The actual stability of tape speed obtained by this method is $\pm 1.5$ parts in $10^{4}$ over the period of one drum revolution (short-term stability) and better than $\pm 1$ part in $10^{5}$ over the period of a complete picture (long-term stability). Fig. l(b) shows how the timing errors are reduced and the picture quality is improved compared with (a).

# Colonial Broadcasting 

By JOHN W. MURRAY*

BROADCASTING within the colonies, which has been in operation for not much more than ten years, is posing problems that demand a fairly early solution. The fundamental problem is similar to that which led the B.B.C. and other European broadcasting authorities to introduce a v.h.f. service, although the attendant circumstances are in general very different.

Twelve bands ${ }^{1}$ within the high-frequency section of the radio spectrum ( $3-30 \mathrm{Mc} / \mathrm{s}$ ) are allocated for broadcasting; some for certain regions and others world-wide. Allowing a "guard" of $5 \mathrm{kc} / \mathrm{s}$ at both ends of each band, there is a total available bandwidth of $2,800 \mathrm{kc} / \mathrm{s}$, which, assuming $10 \mathrm{kc} / \mathrm{s}$ per channel, provides 280 channels. In order to eliminate "skip" effect close to the transmitting station and to ensure that the greater part of the power radiated is concentrated, as far as possible, within the territory to be covered, it is necessary to arrange for highangle radiation. This means that, in practice, the maximum usable frequency cannot be higher than the critical frequency.

For the broadcasting services under consideration, therefore, it is unlikely even at periods of maximum sunspot activity that frequencies much higher than $10 \mathrm{Mc} / \mathrm{s}$ will be used. The lower limit is imposed by the high noise level below $4 \mathrm{Mc} / \mathrm{s}$. Thus, ignoring all bands above $10 \mathrm{Mc} / \mathrm{s}$, we are left with a total bandwidth for local colonial broadcasting of $1,225 \mathrm{kc} / \mathrm{s}$, providing for only 122 clear channels.

Even supposing that it were possible to achieve complete international agreement on the allocation, clearing and sharing of frequencies (and this, unfortunately, is far from being the case) it is obvious that it is utterly impossible to attempt to accommodate all the required colonial broadcasting stations within the usable part of the h.f. bands. The result of trying to do so is interference on most channels and, moreover, there is no room in the h.f. bands for future expansion. The conclusion to be drawn is that some alternative to the use of short waves must be explored for the efficient internal coverage of individual colonial territories. The obvious alternatives are: medium frequencies and very-high frequencies. The use of medium frequencies, although fairly satisfactory, suffers from two major limitations. First the medium-wave band is relatively uneconomical because of the necessity to use fairly high powers to overcome atmospheric noise, and, secondly, because of the possibility of interference from distant stations after dark. V.H.F. transmissions suffer from neither of these disadvantages.

The use of v.h.f. does, however, suffer from some

* Nigerian Broadcasting Service.
${ }^{1}$ Broadcasting bands $(3-30 \mathrm{Mc} / \mathrm{s})$. Those shared with other services are marked with an asterisk. 3.2-3.4*; 3.9-4.0*; 4.75$4.995^{\star} ; 5.005-5.06^{*} ; 5.95-6.20 ; 7.1-7.3^{\star} ; 9.5-9.775$; 11.7$11.975 ; 15.1-15.45 ; 17.7-17.9 ; 21.45-21.75 ; 25.6-26.1$.
important limitations. The first is the lack of cheap receivers, whether f.m. or a.m., battery- or mainsoperated. Clearly, if a decision is made in favour of the use of v.h.f. broadcasting in under-developed countries, the change-over from short-wave broadcasting will have to be spread over a period of perhaps five to ten years. It would be preferable, therefore, for a "general purpose" receiver to be made available, providing for reception on short and medium waves as well as on the v.h.f. bands. This would be desirable, in any case, so as to make it possible for listeners to tune to overseas stations as well as to their own domestic v.h.f. channels. At the present time such a receiver would not cost less than $£ 30$ and, while this might find a fairly ready sale in towns, the rural inhabitants with a much lower income would find it quite impossible to afford such a sum. It is thought that a receiver of the type envisaged could be made available much more cheaply than at present if it were possible for several colonial territories to co-operate so that an initial order of not less than 30,000 receivers could be placed with a single manufacturer. Preliminary discussions have already taken place on this subject with one manufacturer. The servicing of the sort of "combination" receiver suggested might conceivably present major difficulties, particularly if it were decided to use a frequency-modulated v.h.f. system. More complicated test equipment and much greater knowledge and skill would be required to service such receivers compared with the "rule-of-thumb" servicing of a medium-wave or even a medium- and short-wave receiver. A further limitation to the use of both v.h.f. and m.f. broadcasting, particularly in large colonial territories without reliable internal telephone communications, is the problem of expense in supplying programmes to each transmitter. Granted that much could be done to solve this by careful siting and design of aerials (more particularly with v.h.f.), so that some transmitters would also form part of the programme distribution chain.

To summarize, it is certain that there can be little expansion of domestic broadcasting in colonial territories on the h.f. bands, whether by an increase in the number of channels or of the powers used. The alternatives are v.h.f. or m.f. Transmitters for v.h.f. might be considered to have greater range for less power than m.f. transmitters, because of the noise limitation on the m.f. band, and, therefore, be cheaper to run. On the other hand, the use of v.h.f. poses some major problems because of the necessity to provide and service much more complicated receivers. I think that this whole problem of the choice of method of coverage is fundamental to broadcasting in every colonial territory, and that discussions should be held, in the near future, between British radio manufacturers and technical representatives of every colonial territory concerned.


## Modifying Upper-Sidehand Television Receivers

Alterations to T.R.F. Sets for Reception

of the Crystal Palace Transmissions

By G. J. CONWAY

ANUMBER of t.r.f. television receivers built to receive the transmisssions from Alexandra Palace were designed to accept the upper vision sidebands and are now giving unsatisfactory reception of the Crystal Palace transmissions which have a vestigial upper sideband (up to $750 \mathrm{kc} / \mathrm{s}$ ). This article describes the modifications which are necessary to obtain good results from such a receiver.

The original decision to tune the vision receiver to the upper sideband was made in order to reduce, as far as possible, breakthrough of sound signals into the vision receiver and of vision signals into the sound receiver. The outermost lower vision sideband is only $0.5 \mathrm{Mc} / \mathrm{s}$ from the sound carrier frequency and, in a television receiver designed to accept the lower or both sidebands, the response of the vision r.f. circuits must fall away sharply (by say 40 dB ) in this frequency range. Such a response requires the use of at least two sound rejector circuits in the vision receiver. This problem is considerably eased by tuning the vision r.f. circuits to the upper sideband because the separation becomes $3.5 \mathrm{Mc} / \mathrm{s}$, so large that one rejector is usually adequate.

The receiver which was modified was an Alba Model T432, which may be taken as typical of receivers of the upper-sideband type. A simplified
circuit diagram of the r.f. circuits of this receiver before alteration is given in Fig. 1. Of the r.f. circuits $L_{1} L_{2}, L_{3} L_{4}, L_{8} L_{8}, L_{7} L_{8}$ and $L_{8} L_{10}$ the last is a true bandpass double-tuned circuit but the others are, in effect, single-tuned circuits with a single tuning slug. The top-end capacitors $\mathrm{C}_{3}, \mathrm{C}_{2}$ and $\mathrm{C}_{3}$ are 100 pF each which has negligible reactance at the operating frequency and hence $L_{3} L_{s}$ and $L_{7}$ behave as r.f. chokes. Originally these circuits were stagger-tuned to give a passband covering the range 45 to $48 \mathrm{Mc} / \mathrm{s}$, the alignment frequencies being indicated on the circuit diagram. These circuits were retuned by adjustment of the dust-iron slugs to give a passband from 42 to $45 \mathrm{Mc} / \mathrm{s}$. The alignment frequencies are indicated in Fig. 2 and it was necessary to add 5 pF of capacitance to certain of the circuits where the slug adjustment gave insufficient inductance. The passband was checked by a voltmeter connected across $\mathrm{R}_{1}$, the vision detector load, the readings being noted whilst a signal-generator output was swept over the frequency range.

After realignment the picture detail was acceptable but, as expected, there was severe breakthrough of sound on vision. The breakthrough of vision signals into the sound receiver was not as objectionable as expected and could be eliminated with some sacrifice in picture detail by careful alignment of the common r.f. stages. The protection due to the sound rejector circuit $\mathrm{L}_{11}$. was inadequate and a second ( $\mathrm{L}_{12}$ in Fig. 2) of similar design was added to $\mathrm{V}_{4}$ anode. The details of the inductor used are given in the Appendix. The trap is adjusted by means of $C_{5}$ to give minimum output at $41.5 \mathrm{Mc} / \mathrm{s}$ at the vision detector.
With the addition of the second sound trap, the receiver performed much better, there being no horizontal bars across the picture due to sound break-

through. However, there was slight evidence of vision signals breaking into the sound receiver, particularly near maximum or minimum setting of the contrast control. Moreover, there was also visible on the screen a very fine stationary pattern ultimately traced to inter-modulation between the vision and sound carriers. The non-linearity causing this pattern occurred at $V_{3}$ grid due to the large sound signal there. The realignment of the common r.f. circuits at a lower frequency had brought about a considerable increase in sound gain and the signal at $\mathrm{V}_{3}$ grid was large enough to cause cross-modulation. To reduce the sound signal at $\mathrm{V}_{3}$ grid the sound take-
off point was transferred from $V_{3}$ to $V_{2}$ anode, which still left more than adequate sound gain but eliminated the sound patterning and a slight tendency towards instability in the sound receiver. A small capacitance ( 3.3 pF ) was required at $\mathrm{V}_{3}$ anode to make up for the loss due to removal of the sound rejector. After realignment the performance of the receiver was satisfactory.

## APPENDIX

## Details of Sound Rejector Coil $\mathbf{L}_{12}$

10 turns of No. 18 s.w.g. tinned copper wire wound as a self-supporting air-spaced coil of $\frac{1}{4}$ in internal diameter occupying a length of lin.

## COMMERCIAL LITERATURE

Ionospheric Scatter.-Transmitter of 40 kW output, covering $35-55 \mathrm{Mc} / \mathrm{s}$, for frequency shift keying of $4-9 \mathrm{kc} / \mathrm{s}$. Duplicate r.f. amplifiers, normally operating in parallel, for continuity of service. Double diversity receiver, $30-60 \mathrm{Mc} / \mathrm{s}$, with triple frequency changing, using separate "mark" and "space" receiving chains. Leaflet from Marconi's W.T. Company, Marconi House, Chel:nsford, Essex.

Communications Receiver, $15-45 \mathrm{kc} / \mathrm{s}$ and $100 \mathrm{kc} / \mathrm{s}-30 \mathrm{Mc} / \mathrm{s}$, with exceptionally long tuning scale of about 4 ft for each of seven ranges. Has incremental tuning control over constant $100-\mathrm{kc} / \mathrm{s}$ range and incorporates crystal calibrator giving $100-\mathrm{kc} / \mathrm{s}$ check points throughout frequency range. Specification on a leaflet from Airmec, High Wycombe, Bucks.
Television Waveform Generator; valve-voltmeter; audio frequency meter; power supply units; and television and f.m. pre-amplifiers and distribution amplifiers. Illustrated catalogue from Channel Electronic Industries, Dunstan Road Estate, Burnham-on-Sea, Somerset. Also a leaflet on a portable Moisture Meter with a battery life giving about 10,000 readings before replacement.
Closed Circuit Television Equipment for educational and demonstration purposes. Comprises miniature camera, using $16-\mathrm{mm}$ lenses, portable control equipment using plugin units, and direct-viewing monitor. Also two projection equipments available for larger audiences. Leaflets from General Precision Laboratory, 431, Fifth Avenue, New York 16, N.Y., U.S.A.
V.H.F. Aerials for television and f.m.; general aerial accessories; connectors; cables; and a range of test gear. Catalogue, including many items of British manufacture, from Rudolph Schmidt, Gl. Kongevej 64, Copenhagen 5, Denmark
"Transistors for the Experimenter" is the title of a new booklet giving application notes and 17 practical circuits
built around the OC70, OC71 and OC72 Mullard types. Characteristics and circuit configuration are discussed, with sections on large- and small-signal a.f. stages, d.c. stabilization of working point, and oscillator and ewitching circuits. From Mullard, Technical Service Dept., Century House, Shaftesbury Avenue, London, W.C.2.
Impregnation Recording Equipment for use in manufacture of cables, transformers, capacitors, etc. Gives continuous record on paper chart of insulation resistance, self-capacitance (of cables), and temperature and vacuum inside impregnation tank. Leaflet from Addison Electric Company, 10-12, Bosworth Road, London, W. 10.
Germanium Diodes and Transistors. Manual giving full characteristics with curves, application notes and specimen circuits of 18 Telefunken types. A power transistor, OD604, when used in a Class B push-pull audio output stages gives 3.5 W with a 6-V supply. From Telefunken, Mehringdamm 32-34, Berlin S.W.61, Germany.
Power Supply Units, voltage regulated, for $105-125 \mathrm{~V}$, $60 \mathrm{c} / \mathrm{s}$. Have recovery time of less than 50 usec and are suitable for square wave pulsed loading. Voltage range continuously variable, and either positive or negative may be earthed. Leaflet from Kepco Laboratories, 131-138, Sanford Avenue, Flushing 55, N.Y., U.S.A.

Projection Television Screen, $24 \mathrm{in} \times 18 \mathrm{in}$, for use in larger receivers (back projection type) in schools, hotels, clubs, etc. Made of plastic material and has Fresnel and lenticular patterns embossed, as in smaller types. Leaflet from Mullard, Century House, Shaftesbury Avenue, London, W.C.2.
Magnetrons and Travelling-Wave Tubes. Theory of operation, performance characteristics, application notes and techniques for measuring important electrical parameters are given, as well as data on commercially available types, in a booklet from R.C.A. Great Britain, Lincoln Way, Windmill Road, Sunbury, Middlesex, price 4 s 6 d .

## The Great Transistor Chaos

I$T$ is sometimes said that one doesn't have to know how a transistor works in order to be able to use it. In a sense, that is true; but it is as well to consider what sense. For instance, what is meant by " using" a transistor?

In one sense, all the people who are going to buy transistorized portable sets will be using the transistors contained therein. Obviously they will not find it necessary to take a course on the physics of semiconductors in order to be able to do so. Most of them will know nothing whatsoever about how transistors work, and probably would not recognize one if they saw it. In the same sense, users of motor vehicles would include passengers, who in that capacity secure no appreciable advantage from a study of thermodynamics.

But it is rather different for those who drive. Though it is true that very many motorists have only the haziest idea of the means by which their operation of the controls achieves the desired results, this is regrettable; to get the best out of a machine, intelligent understanding is better than memorized rules. It would, however, be pressing the point rather far to insist that drivers ought to be competent in the theory of thermodynamics as well as in practical mechanics. But when we come to a third class -those who design the things-practical mechanics might be enough for turning out some sort of vehicle, but certainly not a good one by present standards.

## Too Many Data?

I assume that Wireless World readers are not content just to listen to transistor radio, or even to control it, but want to apply transistors to their own problems without having to receive word-byword instructions. In my view, it is not enough to regard transistors as little three-terminal boxes having certain characteristics as set forth on the makers' data sheets. Without at least a rough idea of their internal electronics, one is quite likely to ruin them at a very early stage of the practical proceedings. In any case, a willingness to accept them as boxes of mystery is surely rather a poor state of mind?

Hence my recent efforts* to impart the aforesaid rough idea.

The next step is to learn what the makers' data mean. And that is where trouble really begins. If, when studying the electronic side of the subject, one finds the Fermi levels and Pauli exclusion principle too much, one can always give up; such knowledge, though desirable, is far from necessary for ordinary purposes. But if one is in a state of confusion about transistor characteristics, that state is bound to affect everything from there on.

The full extent of the difficulty may not appear at first sight. After all, having learnt what transistors are, one is prepared for their characteristics to be a little more complicated than a valve's. Looking at the first data sheet we find $\alpha$, the current amplification. That is all right; we know that a transistor is basically a current amplifier, so it is natural to have this $\alpha$ in place of a valve's $\mu$. Then instead of

[^6]the familiar $r_{a}$ there are four resistances: $r_{e}, r_{b}$, $r_{c}$ and $r_{m}$. Well, we have been prepared for the fact that the input electrode of a transistor, unlike the biased grid of a valve, is highly conductive, so it is natural for there to be a $r_{b}$ (presumably base resistance) or-and possibly and- $r_{e}$ (emitter resistance) as well as $r_{c}$ (collector resistance). But what is $r_{m}$ ? Something analogous to $1 / g_{m}$ ? Certainly all these resistances are going to make transistor calculations more complicated than valve calculations. And of course there are the alternative methods of connecting it, resulting in widely different amplification factors such as $\alpha$ and $\alpha^{\prime}$. Come to think of it, though, the effective $\mu$ of a valve is also altered very much if the valve is connected as a cathode follower instead of with the more usual earthed cathode; so that is no new complication.

Thus we may comfort ourselves.
Then we look up another data sheet and find a set of completely mysterious quantities called $h_{11}, h_{22}, h_{12}$ and $h_{21}$, to say nothing of $h_{11}{ }^{\prime}, h_{22}{ }^{\prime}, h_{12}{ }^{\prime}$ and $h_{21}$. And that is far from all. We look up a book to get a bit of know-how, and discover that it prefers to deal in $r_{11}, r_{22}, r_{12}$ and $r_{21}$, which are not (as we had dared to hope) just different symbols for the same things. Worse still; in another bookor perhaps the same one-we encounter $g_{11}, g_{22}, g_{12}$ and $g_{21}$. And we learn that the foregoing $r$ and $g$ groups apply only at low frequencies and really ought to be the "complex" quantities $z$ and $y$. And another book, while using the known $r_{e}, r_{b}$, $r_{c}$ and $r_{m}$, also throws in $r_{d}$. And while one technical periodical extols the merits of $r_{i n}$ and $r_{\text {out }}$ in place of $r_{12}$ and $r_{21}$, another advocates a " reverse current amplification," $\beta$. And when we come to equivalent circuits for the transistor (to compare with the two alternative kinds we know for the valve) we find that one book alone $\dagger$ shows dozens of different varieties, merely as samples of many more!

Before you instruct your newsagent to stop the Wireless World and decide to seek admission to the Royal Horticultural Society instead of the I.E.E., may I prevail upon you to do nothing rash. I entirely agree with you that the situation just outlined is one which would excuse, if not justify, a breach of the peace or other departure from normal civilized behaviour. Nor am I able to offer any hope that its obscurities can be dispelled by a few crisp words of explanation. But although there is something to be said for every one of these systems, in the course of time the principle of survival of the fittest will presumably operate; and meanwhile there exist tables for "translating" transistor data from one system to another.

I feel we would be better equipped to find our way through this rank technical jungle if we turned back to review what was done with the valve. This is the approach adopted by W. T. Cocking in his excellent series "Transistor Equivalent Circuits," in the July to October issues last year. Part 1 of that series, in fact, is devoted entirely to the valve; and I recommend you to read (or re-read) it before going any further-even though he did adopt the opposite

[^7]convention to the one I favour $\ddagger$ with regard to the direction of anode current. He had a good reason for doing so, because instead of beginning with the usual "small signal" or a.c. equivalent circuit (which might have puzzled beginners unused to the idea of neglecting such things as h.t. and grid bias sources) he started from the static or d.c. characteristics and thereby was logically almost obliged to decide on the direction of the anode current as being the same as the d.c. component supplied by the h.t. source, even though that source forms no part of the purely a.c. equivalent circuit and its intrusion conflicts with the usual convention of reckoning voltages relative to the cathode. Here, for brevity, I must assume that everyone who had not already grasped the idea of omitting the d.c. parts in an equivalent circuit has made good that deficiency by learning from Mr. Cocking.
And since this subject is complicated enough at the best, I am going to restrict it to (1) triodes (threeelectrode valves and transistors) and (2) lowfrequency operation (ignoring electrode capacitances,


Fig: 1. (a) represents an actual valve, with its grid signal voltoge relative to cathode kept ot zero by short-circuiting g to k . (b) is the equivalent circuit of the valve under these conditions; it can be substituted for (a) for purposes of a.c. signal calculations.


Fig. 2. (a) The.essentials (so far as a.c. is concerned) of the most familiar basic type of valve circuit, in which the input (a signal source) is connected between $g$ and $k$, and the output (a load resistance) between a and $k$. (b) The valve equivalent.
hole storage, and other high-frequency effects). It is, of course, assumed that operation of both valves and transistors is confined to nearly linear parts of their characteristic curves and that, owing to appropriate negative bias, valve grid current is negligible.
Because of this absence of grid current, there is only one current path through the valve-between cathode and anode. The current through this path depends on the voltage across it, just as with any resistor. It also depends on the grid voltage relative to the cathode, but if we wish we can exclude that from the problem by short-circuiting it to cathode, as in Fig. 1(a). So far as a.c. is concerned, the valve so connected behaves as a resistor of approximately constant resistance, which is customarily denoted by the symbol $r_{a}$. The equivalent circuit of the valve is therefore as shown at (b). The value of $r_{a}$ is invariably included among valve data. It can be measured in the same way as any other a.c. resistance, if provision is made for the appropriate d.c. through the valve. Alternatively its value can be derived from the slope of the anode-current/anode-voltage characteristic curves within suitable limits; e.g., if an additional 15 V is needed to increase the anode current by 1 mA , then $r_{a}=15 / 0,001=15,000 \Omega$ or $15 \mathrm{k} \Omega$.

## A Familiar "Equivalent",

The next step is to represent what happens when the grid voltage is variable. The means of varying it is shown in the valve circuit diagram, Fig. 2(a), as a generator giving an alternating voltage $v_{k g}$. Note that this symbol signifies the e.m.f. in the direction $k$ to $g$, which means that when it is positive it makes $g$ positive (relative to $k$, which is reckoned as zero potential). Again, the valve-the part of the diagram within the dotted line-can for purposes of calculation be replaced by the equivalent circuit, Fig. 2(b). The influence of $v_{k g}$ is represented here by a generator in series with $r_{a}$ delivering a voltage - $\mu v_{k g}$; where $\mu$ is the universally accepted symbol for the valve's voltage amplification factor, equal to the number of anode volts that have to be applied to offset minus one grid volt. This too can be measured, or derived from sets of characteristic curves which show the effects on anode current of both anode and grid voltage. Because there is no grid current the grid terminal in the equivalent diagram is not connected internally, and indeed is often omitted, except in diagrams where interelectrode capacitances are taken into account.
A valve is normally connected at two points to an input and at two to an output, making four connecting points altogether; but as there are only three electrodes one electrode must be common to both input and output, being joined to them either directly or through negligible a.c. impedance. In each case it is the low-potential terminal of the pair that is "commoned," and these terminals are often kept at their constant potential by earthing, as shown in Fig. 2(a). The basic valve circuit "configurations," as they are termed in America, are distinguished by the electrode that is common to both input and output. With a triode there are therefore three of them, called by true Britons "earthed-cathode,"
" earthed-anode" (or, for other reasons, " cathode follower ") and "earthed-grid." Americans, in accordance with national custom and therefore quite logically, say "grounded" instead of "earthed." Certain persons in this country, contrary to national custom and even their own custom in other contexts and therefore quite illogically, also say "grounded". In view of this situation, which almost led to an unedifying fight between a contributor and the Editor, § I have decided to fall in with a growing practice on both sides of the Atlantic and abandon both words in favour of "common." Not only is this free from any taint of nationalism but it more truly describes the condition, earthing not being the really essential feature. The earth symbol in Fig 2(a) and other diagrams is there mainly to call attention to which is the common low-potential electrode.

## The Preferred Circuit

Theoretically, one could argue that there are six basic configurations. For each of the three choices of common terminal there is a choice of which of the other two is to be input and which output. But in each case only one of these gives amplification, so the other is not usually counted.

With valves, the common-cathode arrangement is the original and by far the most commonly used, so the accepted practice has always been to measure valve characteristics in this condition and to reckon voltages relative to cathode-hence my strong preference for the convention shown in Fig. 2(b), where the artificial anode generator gives minus $\mu v_{l g}$ volts relative to $k$, and the positive direction of signal current is from $k$ to $a$ as indicated by the arrow, opposite to the steady d.c. component. Otherwise one has to suppose that this generator acts from $a$ to $k$, contrary to the convention of regarding $k$ as the starting point.

But that is by the way. The important thing just now is the fact that there is, and always has been, no doubt at all abjut which of the three configurations is assumed when mentioning the two essential triode "parameters," as they are called: $r_{a}$ and $\mu$. If the first use of triodes had been as cathode followers it would have been quite different: as we shall soon see, $\mu$ would always have been less than 1, and $r_{a}$ much less than we are used to, and conversion formulae would have been necessary to derive the values applicable in Fig. 2(b).

Another important point is that, thanks to the nonconducting grid path, there are only these two basic parameters, $r_{a}$ and $\mu$. But what, you may say, about the mutual conductance $g_{m}$ ? Although $g_{m}$ is nearly always included among valve data it does not make a third basic quantity, because if the first two are known it is known too, being equal to $\mu / r_{a}$. Moreover it is not needed at all in the well-known form of equivalent circuit we have seen-Fig. 2(b). But as an alternative to this equivalent, with its voltage generator in series with $r_{a}$, there is another form, with a current generator $\|_{\text {in }}$ parallel with $r_{a}$ Fig. 3. One can show (as I did in the April 1951 issue-" That Other Valve Equivalent ") that it

[^8]

Left: Fig. 3. An alternative to the voltage generator in series with $r_{i}$ is a current generator in parallel.

Below: Fig. 4. (a) the cathode follower or common-anode circuit. (b) The circuit equivalent of the volve, adapted to be in terms of the signal e.m.f. $\mathrm{v}_{\mathrm{ag}}$.

(a)
(b)
always gives exactly the same results. But it is sometimes more convenient; especially with tetrodes and pentodes, which are covered by the triode equivalent circuits so long as the extra electrodes are tied down to constant potentials. Their $\mu$ and $r_{a}$ are so high that Fig. 3 is the more natural choice; in fact, if the load impedance is very low compared with $r_{a}, r_{a}$ can be omitted without making much difference to the calculations.

Since in Fig. 3 the load impedance comes directly in parallel with $r_{a}$, it is more logical to think of it as a load admittance, which can be added directly to the anode conductance $g_{a}\left(=1 / r_{a}\right)$ to give the total admittance fed by the current generator. I just mention this to show that although (within the restrictions we have assumed) there are only two basic valve parameters, one can make for oneself quite a wide choice as to which two, some being more convenient for some purposes and others for others. Fortunately the relationships between them are simple, so that (as we have seen) it is quite easy to transfer from $\mu$ to $g_{m}$ or from $r_{a}$ to $g_{a}$. And there is no risk of the situation getting out of hand; there are clear reasons for quoting primarily $\mu$ and $r_{a}$ for triodes and $g_{m}$ and $r_{a}$ or $g_{a}$ for tetrodes and pentodes, and nobody is in any doubt about the common-cathode being the configuration to which they apply.

But the other two configurations are used quite frequently, so we must see what happens about their equivalent circuits. Let us take the common-anode circuit or cathode follower first. Fig. $4(\mathrm{a})$ is the circuit diagram in its simplest form. The valve could of course be drawn with its terminals in the same positions as in Fig. 2(a), but the diagram would then be difficult to recognize as a cathode follower, so I have drawn it in the customary manner with the low-potential terminal brought down to the foot. In a practical circuit, of course, there would be a blocking capacitor between this and the anode so as not to short-circuit the h.t., but to signals there is (or should be) effectively a short-circuit.

Fig. 2(b), let us take care to note, is still a valid circuit equivalent of the valve, because within its dotted line the valve itself is no different for being in a cathode-follower circuit or any other, provided it gets its proper ration of d.c. But Fig. 2(b) is not a convenient equivalent now, because it is based on $v_{k g}$, which is not shown in Fig. 4(a). Moreover, although true, it is misleading, because $v_{k g}$ is affected by any external voltage applied across $a k$, which causes the equivalent internal generator to behave in such a way as to make the resistance seem lower than $r_{a}$.

So the normal practice is to recalculate the resistance and generator e.m.f. in terms of the external signal voltage $v_{a \theta}$, and the result is shown in Fig. $4(b)$. Note that although the generator arrow is still pointing upwards the path itself has been reversed to match diagram (a), since the low-potential terminal of the output is now $a$ instead of $k$. This reversal cancels out the minus sign that was needed in Fig. 2(b), and means that the output voltage has the same polarity as the input. Note that the equivalent generator e.m.f. and resistance are in terms of the common-cathode parameters $r_{a}$ and $\mu$, but that is simply a tribute to the relative importance of the common-cathode configuration. There is nothing to stop anyone measuring valves' internal resistances and amplification factors with the common-anode connection and inventing symbols for them, say $r_{a}^{\prime}$ and $\mu^{\prime}$. However, as these would be relatively seldom used and their equivalents in Fig. 4(b) are not difficult to remember, there would be no great point in the exercise, so they are not included in published valve data.

Just in case there are any lingering doubts about the two equivalent circuits, Figs. 2(b) and 4(b), being equivalent to one another, let us take a simple numerical example. Suppose the valve used in the cathode follower circuit, Fig. 4(a), has a $\mu$ of 24 and $r_{a}$ of $10 \mathrm{k} \Omega$, and that the load resistance RI is
know to be 2V. It is therefore -48 V . But this apparent increase in voltage is exactly offset by the higher internal resistance, $10 \mathrm{k} \Omega$. So the output is again 8 V -Fig. 5(b). The minus sign for the generator voltage doesn't make any difference, because in this equivalent circuit the generator voltage is reckoned from cathode. The output voltage is still in the same direction as with the other equivalent, reckoned from anode. Although the second method of calculation gives the same answer as the first, it is less convenient, because we had to know the answer in order to find it. The first method, on the other hand, gives it direct from the gross input, $v_{a g}$.

## More Like a Transistor

Lastly, the common-grid circuit, Fig. 6(a). Though not drawn in quite the usual way, it should be easily recognizable. Again, Fig. 2(b) is still a true equivalent, but again it is misleading. Unlike the other two circuits, with this one the load is connected between anode and grid, which means that the signal source is in series with the currentcarrying path through the valve, and so the impedance of this source comes into the problem and must be shown. In Fig. 6(a) it is the resistance $\mathrm{R}_{\mathrm{s}}$ : Because of it, the e.m.f. of the source is not the same as the voltage between $k$ and $g$, so cannot appropriately be called $v_{k g}$. I have, therefore, marked it $v_{s}$; and as that label does not indicate its direction an arrow is shown instead, to make quite clear that this time, because $g$ is the low-potential terminal, the input is reversed. The high-potential terminal of the output is $a$, the same as in Fig. 2, so again there is one reversal to cancel out the minus, giving the output the same polarity as the input.

Although the valve equivalent generator voltage, $\mu v_{s}$, makes no allowance for the loss in $\mathbf{R}_{s}$, that is all right because the loss is exactly allowed for by

(a)

(b)

Left: Fig. 5. Numerical calculation of a particular cathode follower, using (a) the valve equivalent devlsed especially for cathode followers and shown in Fig. 4(b), and (b) the common-cathode equivalent shown in Fig. 2 (b).

Below: Fig. 6. (a) Common-grid circuit, showing the signal source resistance $\mathrm{R}_{\mathrm{s}}$ because this affects the signal voltage actuolly reaching the valve. (b) Equivalent circuit of the valve, in terms of the source e.m.f. $\mathbf{v}_{3}$.
$2 \mathrm{k} \Omega$ and the input ( $v_{a g}$ ) is 10 V . Then, using the Fig. 4(a) equivalent circuit we calculate the internal resistance as $10 / 25=0.4 \mathrm{k} \Omega$, and the internal generator voltage as $24 / 25 \times 10=9.6 \mathrm{~V}$. This 9.6 V is distributed between the internal resistance and $\mathrm{R}_{\mathrm{L}}$ in proportion to their resistances, to $\mathrm{R}_{\mathrm{L}}$ 's share works out at $9.6 \times 2 /(2+0.4)=8 \mathrm{~V}$, as shown in Fig. 5(a). Note that this makes the net iniput to the valve-from cathode to grid- 2 V , being the difference between the gross input of 10 V and the 8 V output fed back as a result of the signal being connected from anode.

Now make the same calculation using the other equivalent, Fig. 2(b). Here the internal generator voltage is $-\mu$ times the net input $v_{k g}$, which we now



Fig. 7. Circuit diagram for calculation purposes, obtained by substituting fig. 6(b) in Fig. 6(a).
the addition of $\mu \mathrm{R}_{s}$ to $r_{a}$ in the internal resistance. The reason for the multiplier $\mu$ is that the voltage drop across $\mathbf{R}_{s}$ •affects the voltage between cathode and grid, and this effect is amplified by the valve.

The fact that the input and output circuits are connected directly in series with one another means that the resistance "seen" by each when it is connected depends on the resistance of the otherthat is to say, something outside the valve-and makes this configuration more like a transistor circuit than the other two. This point of resemblance to the transistor is interesting, and to get ourselves in training for transistor circuits let us reckon what the input and output resistances are. By " input resistance " I don't mean $R_{s}$, the resistance of the input circuit, but the resistance the input circuit works into between $g$ and $k$.

## Input and Output Interaction

Take the output resistance first. Substitute Fig. 6 (b) for the valve in (a), to give Fig. 7, and add up the resistances from $a$ to $g$ through the valve and input circuit. There is $r_{a}+\mu \mathrm{R}_{s}$ and $\mathrm{R}_{s}$, so the total is $r_{a}+(\mu+1) \mathrm{R}_{s}$. Clearly that could be a lot higher than $r_{a}$, which is what one gets in the commoncathode circuit.

Now the input circuit. One might suppose that it sees $\mathbf{R}_{\mathbf{L}}+r_{a}+\mu \mathbf{R}_{\mathbf{s}}$. But just as in Fig. 4 a voltage applied between the output terminals gets to the grid and causes an amplified current to flow through the source of the voltage, giving it the impression that it is feeding a much lower resistance than $r_{a}$, so in Fig. 6 the source of input voltage drives a current through itself, since it is in series with the output circuit. That current is equal to the total e.m.f. acting (which Fig. 7 shows to be $(\mu+1) v_{s}$ ) divided by the total resistance, $\mathbf{R}_{\mathbf{L}}+r_{a}+(\mu+1) \mathbf{R}_{s}$ :

$$
i=\frac{(\mu+1) v_{s}}{\mathbf{R}_{\mathrm{L}}+r_{a}+(\mu+1) \mathbf{R}_{s}}
$$

But the signal source, knowing nothing about the internal generator, imagines its own e.m.f. is causing all this current, and judges the input resistance to be $v_{\mathrm{s}} / i$, less its own internal resistance $\mathrm{R}_{s}$ :

$$
\begin{aligned}
\mathbf{R}_{i n} & =\frac{\mathbf{R}_{\mathbf{L}}+r_{a}+(\mu+1) \mathbf{R}_{s}}{\mu+1}-\mathbf{R}_{s} \\
& =\frac{\mathbf{R}_{\mathbf{L}}+r_{a}}{\mu+1}
\end{aligned}
$$

That is very low-much lower than the output resistance-so this valve circuit resembles a transistor amplifier quite closely. But there is still the absence of grid current to simplify the equivalent circuit. It may help us to tackle the more difficult problem
next month if we sum up the points about valves that we should keep in mind:
(1) There are three possible two-terminal paths through a triode (including a tetrode or pentode if the extra grids are tied down to other electrodes)grid to cathode, grid to anode, and cathode to anode -but under working conditions there is no grid current, so only the last of these three is a currentcarrying path and, therefore, is the only one to appear in any equivalent circuit diagram.
(2) Consequently there are only two basic parameters -one to specify the resistance of the single currentcarrying path, and the other to specify the control over it possessed by the grid.
(3) Because the valve has only three terminals, the input and output pairs of terminals must have one in common.
(4) Depending on which of the three valve electrodes is connected to the common input-output terminal, there are three "configurations" or basic arrangements of a valve in its circuitry.
(5) Of these three configurations, the commoncathode one is by far the most important, so is universally accepted as the choice for measuring the valve parameters. The standard symbols for these parameters- $r_{a}$ and $\mu$-are therefore well understood to refer to common-cathode connection.
(6) The parameters for the other two configurations are usually given in terms of the common-cathode symbols; no special symbols for them have been officially appointed.
(7) Although any equivalent that may be devised to represent the valve is valid for any circuit configuration in which the valve may be placed, for practical convenience the equivalent is devised so as to be in terms of the quantities that are known in that particular configuration.
(8) For some purposes it is convenient to use derivatives from the two basic parameters, especially $\mu / r_{a}$ and $1 / r_{a}$. These are sufficiently important to have been given the special symbols $g_{m}$ and $g_{a}$ respectively; and often $g_{m}$ is the quantity to be directly measured, rather than $\mu$.
(9) Because there are two elements in each equivalent circuit diagram there are two varieties of it-series and parallel.

It will be by comparison with these valve facts that we will attempt to bring some order out of the transistor chaos.

## STERIOPHONIC HEAD

A COMBINED playback and recording head for twinchannel stereophonic tape records has been developed by Truvox, Ltd., 15, Lyon Road, Harrow, Middlesex. It


Truvox TR2049 stereophonic recording / playback head is available as a separate unit and
costs $£ 1410$ s. costs $£ 14$ 10s.
The windings are of highimpedance type ( $50 \mathrm{k} \Omega$ at $10 \mathrm{kc} / \mathrm{s}$ ), and are stated to have a frequency range, with suitable amplifiers, of $50 \mathrm{c} / \mathrm{s}$ to $15 \mathrm{kc} / \mathrm{s}$. The gap width is 0.00025 in.

For recording`a bias voltage of 120 V , $5 . \mathrm{m} . \mathrm{s}$. is recommended with a recording current of approximately 0.1 mA . The output on playback is of the order of 1 to 3 mV .


Fig. I. Map showing relationships between the Belgian language groups and the television stations of neighbouring countries providing a useful signal for Belgian viewers.

TELEVISION engineers visiting Belgium are usually surprised to find that the four-standards problem which we have here must be solved for almost every viewer. The present receiver market, although small by comparison with that in Britain, is expanding rapidly. Even so, the call for simple television sets able to receive one or two stations of similar transmission characteristics is virtually non-existent, a fact which will be more readily understood if one takes into account the psychological factors present.

There are two major languages spoken in this small country of $8,000,000$ inhabitants, French in the south and part of the east, and Flemish in the north and part of the west. In addition, in parts of the south-east, German is spoken, although a large part of the population is able to understand both French and Flemish more or less fluently. A glance at a map, Fig. 1, will therefore indicate the origin of some of the problems. The Flemish living on the French border wish to receive Dutch and German transmissions, while the French-speaking viewers on the Dutch and German borders go to extraordinary lengths to get acceptable pictures from Lille.

One may look hopefully towards the Grand Duchy of Luxembourg, which transmits in French. This is very acceptable to the sparse population of southern Belgium, but most upsetting to the viewers in the Grand Duchy, who normally speak German and tune their sets to the German station situated in the Black Forest. At all events the Luxembourg transmissions are, in fact, intended for France!

It will thus be seen that for a very long time to come the demand for all receivers in Belgium and the Grand Duchy must inevitably be for "fringearea " models able to receive any type of transmission current in Europe except the " $8-\mathrm{Mc} / \mathrm{s}$ channel

## FourwStandards Television

Problems of Receiver Circuit Design

in Belgium

By H. de LAISTRE BANTING, A.M.Brit.I.R.E.


Fig. 2. Elaborate television aerial on a private house in the Avenue Terveuren, Brussels.
C.C.I.R." from the U.S.S.R., which is nevertheless receivable in some parts of Benelux. Furthermore, except for Lopik in Channel 4, Liége in Channel 3 and Antwerp in Channel 2, all stations are in Band III.

The aerial systems are necessarily in keeping with the viewer's determination not to miss anything! Fig. 2 shows one of the more exaggerated private installations in Brussels. Throughout the country, and in the Grand Duchy, one can see 7 to 10 element stacked Yagis mounted on masts well over 100ft high. In fact one of the standard masts sold
is 30 metres high and fitted with a rotary mechanism. This sort of thing is more understandable when one considers the location of many of these people living in the country districts without other forms of entertainment, to which is added the language difficulty already explained above. As elsewhere in Europe, a great deal of Belgian television is to be found in the cafés.

From Table I we see the principal differences between the various standards. It is not often realized abroad that during a Belgian programme the announcer will ask viewers to switch from 819 lines to 625 lines or vice versa. This is because Belgian stations retransmitting a Eurovision programme do not "convert" the number of lines except in the case of programmes originating from the B.B.C. This state of affairs immediately precludes coupling the line timebase frequency switch to the channel selector, which would have seemed the obvious thing to do.

The reader will no doubt have thought of the hum problem, which could be solved only by synchronizing the mains of all neighbouring countries. There is another aspect which is fraught with danger. The daily retransmissions in Belgium of the Paris programme are done by using only the picture (on halfbandwidth), and substituting the Belgian sync waveform. In the early days this was disastrous for the flywheel line timebase. Synchronizing the frame

Table I

| Origin |  |  |  |  | Modulation: |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | sound | vision |
| France ... | 819 | 10.4 | 11.15 | lower | A.M. | pos. |
| $\begin{gathered} \text { Belgium } \\ \text { (Flemish) } \end{gathered}$ | 625 | 4.5 | 5.5 | higher | A.M. | pos. |
| $\begin{aligned} & \text { Belgium } \\ & \text { (Fr. lang.) } \end{aligned}$ | 819 | 4.5 | 5.5 | higher | A.M. | pos. |
| Luxembourg | 819 | 4.5 | 5.5 | higher | A.M. | pos. |
| Holland ... | 625 | 4.5 | 5.5 | higher | F.M. | neg. |
| Germany ... | 625 | 4.5 | 5.5 | higher | F.M. | neg. |

timebase in the presence of heavy impulse interference is especially difficult when using the French transmission direct from Lille, due to the fact that the frame sync waveform consists of a single pulse of $20 \mu \mathrm{sec}$ duration. This has the advantage of giving a somewhat better result on interlace, but the sync is made very much "weaker" by the technique.

## Necessary Compromises

After a little experience of the conditions obtaining in Belgium one is forced to certain conclusions on the need for applying a number of compromises, in order to have reasonably acceptable pictures from all directions.

Due to the uncertainty of the black-level in Belgian transmissions, plus the inherent difficulties encountered in d.c. restoration on negative modulation, it is necessary to abandon the d.c. component. Automatic gain control is universally used, but is never a fully gated system. This attitude seems to be justified on the double basis that the pictures do not suffer visibly on peak-white, and that when one has a fairly large number of fringe transmissions more or less within range the inexpert viewer is not able to cope with a contrast variation depending upon the synchronization of one or other of the timebases. This is particularly important when the viewer is not sure if the transmitter in question is receivable and has no set sequence of operations to apply.

All channel selectors are fitted with coils for the C.C.I.R. Channels 2-11; in the 12th position is the French Channel 8A for Lille. This is "rearranged" by changing the oscillator from the high to the low frequency side of the vision carrier. This keeps the vision i.f. the same for all channels, but gives a second sound i.f. which is $5.65 \mathrm{Mc} / \mathrm{s}$ lower than that for the other channels. The remaining position is normally left without coils and can be used for local f.m. broadcast reception if required. Reference to Table I will make this clear.

The introduction of "transformerless." techniques was greatly complicated here by the existence of three mains voltages, 110,130 and 220 volts. Fortunately one can ignore the d.c. mains but not so the 130 volts, which is very common. This naturally makes all voltage-doubler arrangements unpleasantly complicated, added to which the fluctuation in the mains at various times of the day is quite extrava-

Note.-All Belgian stations change from 625 to 819 lines or vice versa depending upon Eurovision relays.

Fig. 3. Typical arrangement for automatic interference reduction. Oǹ positive modulation, negative-going interference peaks cut off PL83. Earthed-grid stage provides amplified signal of same phase for sync separator. On negative modulation, sync is taken from PL83 anode.



Fig. 4. Use of a second frequency changer for sound, giving a second i.f. of either 11.8 or $7 \mathrm{Mc} / \mathrm{s}$. The buffer stage is switched between 27.75 and $33.4 \mathrm{Mc} / \mathrm{s}$. (Switch positions: 1, France; 2, Belgium and Luxembourg; 3, Belgium and Luxembourg; 4, C.C.I.R.)
gant. One finishes by finding that the saving in cost is very much less than in Britain. In the author's opinion the only simple solution is to design for 220 volts and add an auto-transformer for the lot.

Under the conditions described the simple type of vision interference limiter used in British receivers is quite insufficient. It is usual to start with vision demodulation in the negative sense and then apply the negative-going video signal to a video amplifier which reduces the interference to a little over the maximum modulation. This is very effective but entails the use of an earthed-grid triode stage to amplify the composite video signal across the cathode resistance of the video amplifier without changing its phase, in order to provide a suitable signal for the sync separator (see Fig. 3).

## Major Difficulties

Of the three purely "four-standards" problems, the sound is probably the most untidy. There are three principal systems: the use of two entirely separate sound amplifiers; the switching by capacitance of the frequency of a single amplifier; and the use of a second frequency changer. This last is the most popular because of the opportunity it presents to make use of a relatively low frequency ratio detector for the f.m. on the C.C.I.R. transmissions. The ratio detector is at present the most effective solution for the f.m. since the simpler methods of demodulation result in far too much noise on the majority of receivable transmissions.

The difficulty of this solution is to choose a frequency for the oscillator which will be the least troublesome from the point of view of r.f. or i.f. harmonics. To reduce these bad effects it is found necessary, in addition to the more obvious precautions, to use a buffer amplifier between the take-off point in the vision i.f. and the second frequency changer (see Fig. 4). This buffer must be switched in frequency for the transmissions from France because, as has already been explained, the sound


Fig. 5. Switching of c.r.t. modulation. Also, c.r.t. onode voltage compensation to maintain constant brightness when changing from 625 to 819 lines. (Switch positions: same as in Fig. 4.)
i.f. output from the tuner under these conditions is $5.65 \mathrm{Mc} / \mathrm{s}$ lower than normal. The second sound i.f. is then made either $7 \mathrm{Mc} / \mathrm{s}$ or $11.8 \mathrm{Mc} / \mathrm{s}$. Once again the question of interference limiting is of great importance, and a simple limiter will not suffice.
Coming now to the video circuits, reference has already been made to the fact that many manufacturers prefer to use permanent negative demodulation. In this case the positive/negative switching is achieved either by switching in or out a second low-gain video stage, or more usually by switching the modulation of the c.r. tube from the grid to the cathode (see Fig. 5). In other cases the detec-

process technically by either converting all Eurovision to the local standard, or, better still, abandoning the use of two different standards-if this can be done without provoking a political crisis.

A receiver with a 17 in tube costs on the average £100, but more precise figures are difficult to arrive at. A customer does not normally pay the price on the ticket; he will usually insist on a rebate of $10-20 \%$ and the dealer will have to pay at least a proportion of the sales tax, which totals $13 \%$. On the other hand, the customer seems quite happy to pay from £25 to £50 for a very
tion is switched by germanium diodes, by switching the heaters of a double triode where the two "diodes" are permanently connected, or by various systems of biased diodes, which introduce decoupling difficulties. In the cases of switching the detection, it is normal to a.c. couple to the video stage, otherwise additional switching would be required to change the bias of the valve.

Although there exist receivers which employ a special amplifier to preserve the large bandwidth on French transmissions, these do not represent a significant proportion of the market. All wellknown receivers use a $4.5-\mathrm{Mc} / \mathrm{s}$ vision i.f. for all signals. Of course, the channel selector must preserve the complete bandwidth in order to provide the sound signal. For this reason the gain on the Lille channel is about half that on the other channels.

Finally, there are the practical difficulties encountered in switching the line timebase speed. This naturally gives rise to a change in the e.h.t. and the h.t. boost rail, causing changes in height, width and picture brightness. These all have to be corrected, entailing a four-circuit switch. It is normal practice to adjust picture width for 819 lines and to switch a series resistance into the h.t. supply of the line output valve to give the same width on 625 lines (see Fig. 6). Vertical compensation will depend upon whether the normal h.t. or the boost rail is used for the frame timebase. The change in brightness is mostly due to variation of the anode voltage of the c.r. tube, and this is corrected by shortcircuiting a section of a fixed potentiometer to give the desired correction (see Fig. 5).

Regarding the general prospects in Belgium, one may be consoled by the belief that conditions will tend to become simplified-at least so long as one does not take too seriously the recent German statements on their intentions for colour television in the near future. In Belgium, too, there is much talk of colour for the 1958 exhibition.

Naturally, as the transmissions improve technically we can hope for increased programme value. This could have the far-reaching effect of reducing interest in the four-standards receiver to the point where simple receivers would be saleable. In fact, the Belgian broadcasting authorities could assist this
ordinary aexial array mounted on his roof.
The desirable evolution of Belgian television towards a reasonably priced, simple piece of apparatus, providing a standard of entertainment high enough to attract the customers, is a definite possibility during the course of the next year or two, if only the broadcasting authorities are not forced to take some precipitous action in colour television as a result of pressure from outside interests.

## CLUB NEWS

Barnsley.-" Specialized frequency meter" is the title of the talk to be given by W. Richardson (G8VX) at the meeting of the Barnsley and District Amateur Radio Club on November 9th. A fortnight later the club will hold an exhibition and demonstration of members' equipment. Meetings begin at 7.0 at the King George Hotel, Peel Street. Sec.: P. Carbutt (G2AFV), 33 Woodstock Road, Barnsley.

Belfast.-At the September meeting of the City of Belfast Y.M.C.A. Radio Club, an American spoke on amateur radio in the U.S.A. The October meeting (17th) will consist of amateur and professional tape recordings. The club meets at the Y.M.C.A., Wellington Place. Sec.: R. J. Boal (G13AXI), 127 Hillman Street, Belfast.

Birmingham.-At the meeting of the Slade Radio Society on November 9th, T. J. Hayward, of the R.A.F. School of Radio, will deal with microwave techniques. Meetings are held at 7.45 at the Church House, High Street, Erdington, Birmingham, 23. Sec.: C. N. Smart, 110 Woolmore Road, Erdington.

Bradford.-The 1956/57 syllabus of the Bradford Amateur Radio Society includes, in addition to a variety of lectures, a number of visits to works, etc., among them Mains Radio Gramophones Limited, Beckside Works, on October 23 rd. Meetings are held on alternate Tuesdays at 7.30 at Cambridge House, 66 Little Horton Lane, Bradford. Sec. F. J. Davies (G3KSS), 39 Pullan Avenue, Eccleshill, Bradford, 2.

Newbury.-W. H. Allen (G2UJ) will speak on "A Ham in peace and war" at the meeting of the Newbury and District Amateur Radio Society on November 9th at 7.30 at Elliott's Canteen, West Street.

Sidcup.-The next meeting of the Cray Valley Radio Club will be held on October 23rd and will comprise an exhibition of members' home-built gear. Meetings are held at 8.0 at the Station Hotel, Sidcup. Sec.: S. W. Coursey (G3JJC), 49, Dulverton Road, New Eltham, London, S.E.9.

# Inexpensive Variable-Slope Filter 

SIMPLE TREBLE ATTENUATION CIRCUIT

FOR USE IN HIGH-QUALITY AMPLIFIERS

By D. M. LEAKEY, B.Sc.(Eng.), Grad. I.E.E., A.C.G.I.

I$T$ is now generally accepted that in wide-band high-quality amplifiers it is almost essential to provide an adjustable means of limiting the high frequency response of the system in order that the best results may be obtained from the wide range of programme sources available. For the best results this high frequency cut-off should be both adjustable in frequency and in the rate of cut-off. Resistive networks as part of a feedback system can be used for this purpose. To obtain the necessary maximum rate of cut-off however it is necessary to employ twin-T resistance capacitance networks which unfortunately require close tolerance components and relatively complex switching arrangements. The filter to be described produces the same results as the feedback twin-T network but avoids the two above difficulties.

The use of an inductor might be regarded by many as undesirable but in this application the foundations for this opinion are in general almost groundless. Hum pick-up can be troublesome but, especially with relatively simple hum-bucking arrangements, in most applications can be made negligible. Harmonic distortion originating in the inductance has been found to be negligible providing the unit is used where the signal level is not too high. As a very general guide, with normal small iron-cored inductors, this level is about two volts. Similarly, ringing in the circuit is often put forward as a fault, but in general it can be shown that this is no worse than when using R-C networks producing the same frequency response.

The basic circuit for the filter is the single section, resistance terminated, constant- $k$ filter section as shown in Fig. 1. This provides a response as shown in Fig. 2, the slope of cut-off being about $20 \mathrm{~dB} /$ octave. A steeper slope than this would be desirable and this can be achieved by shunting the inductor with a small capacitor so producing a simple m -derived section. The resulting response is shown in Fig. 3 from which will be seen that although a steeper cut-off is achieved the response rises again after the resonant point. In order to limit this rise so that it only approaches to within about 25 dB of the mid-frequency level it is necessary to limit the value of $\mathrm{C}_{3}$ to about one-tenth of $\mathrm{C}_{1}$ or $\mathrm{C}_{2}$. With this value an initial cut-off slope of


Fig. 1, Basic constant-k filter section.

This article describes a simple m-derived single LC filter section which provides a variable slope of cut-off between about $40 \mathrm{~dB} / 0$ octave and $6 \mathrm{~dB} /$ octave at any selected roll-off frequency. It requires no close tolerance components and requires no initial adjustment.


Fig. 2. Typical response of the circuit of Fig. I.



Fig. 3. A steeper slope is obtained when the series inductance is shunted by a capacitance.
$40 \mathrm{~dB} /$ octave can be achieved whilst the response above the cut-off frequency never returns to more than within 25 dB of the mid-frequency level. A return level of 25 dB has been chosen only as an arbitrary figure which in practice has been found to be satisfactory. Actually when the m-derivation was included (i.e., the connection of $\mathrm{C}_{3}$ across the inductor) the initial values of $\mathrm{C}_{1}, \mathrm{C}_{2}$ and L should have been slightly modified, but due to the fact that $\mathrm{C}_{3}$ is only one-tenth of $\mathrm{C}_{1}$ or $\mathrm{C}_{2}$ the error produced by leaving them at their original values is, for this application, negligible.

To achieve a variable slope of cut-off it is necessary to vary the Q of the tuned circuit formed by L and $\mathrm{C}_{3}$. The result of this is shown in Fig. 4 where a variable resistor $\mathrm{R}_{3}$ is placed across L . As can be seen, the initial cut-off slope can easily be varied between 40 dB /octave and 6 dB /octave whilst leaving the cut-off frequency substantially constant.

In the above circuits a constant-voltage source is shown. This is not necessary since the input resistor $\mathbf{R}$ can be modified to allow for a finite source resistance. Similarly $R_{2}$ can be modified if an external load is placed across the output or $\mathrm{C}_{2}$ modified to allow for any input capacitance of the load.

A practical circuit giving a variable slope of cutoff at about $5 \mathrm{kc} / \mathrm{s}, 7 \mathrm{kc} / \mathrm{s}$ and $10 \mathrm{kc} / \mathrm{s}$ is shown in Fig. 5. It will be noticed that resistors and capacitors are switched, leaving the inductance value constant. This is purely for economic reasons bearing in mind the possible relative price of the components. For smooth slope control the variable resistor should be of the logarithmic law type. The circuit can be fed from any low-impedance source such as a lowimpedance triode, or a pentode with heavy negative feedback. For the $5-\mathrm{kc} / \mathrm{s}$ cut-off condition the resonant dip can be adjusted by means of $\mathrm{C}_{3}$ to occur at $9 \mathrm{kc} / \mathrm{s}$ and so be useful as a whistle filter on medium-wave broadcast reception.

The components in the above circuit are not critical, although at least a $\pm 10 \%$ tolerance on the components is preferable. The one-henry inductance can conveniently wound on a Mullard Ferroxcube La7 core, with about 40 s.w.g. enamelled wire. The experimental coils required about 1,100 turns for 1 henry. In practice however it is preterable to wind on say 1,200 turns, measure the inductance
and remove the correct number of turns to reduce the inductance to 1 H , remembering that the inductance is very nearly proportionally to the square of the number of turns.

Humbucking can be achieved after complete assembly of the core and coil by winding turns over the complete core in reverse to the main winding. The number of turns can be found experimentally by measuring the hum pick-up of the coil when it is near a source of magnetic hum. Only a very few turns are normally required, insufficient to materially alter the inductance from the required one henry.

## APPENDIX: Design of filter

IT is assumed that the known quantities are the required cut-off frequencies and the value of inductance available: L in henrys.
C in farads.
R in ohms.
$f_{\mathrm{c}}=$ cut-off frequency in $\mathrm{c} / \mathrm{s}$.
$\left.\begin{array}{l}\mathrm{C}_{1}=\mathrm{C}_{2}=\frac{1}{2 \pi^{2} f^{2}{ }^{2} \mathrm{~L}} \\ \mathrm{R}_{1}=\mathrm{R}_{2}=\pi f_{\mathrm{L}} \\ \mathrm{R}_{3}=200,000 \mathrm{~L} \text { (approx.) exact value is unimportant. }\end{array}\right\} \begin{aligned} & \text { Choose the nearest } \\ & \text { value for components. } \\ & \text {. }\end{aligned}$
$\mathrm{R}_{3}=200,000 \mathrm{~L}$ (approx.) exact value is unimportant.


Fig. 4. Variation of slope is achieved by altering the Q of the LC circuit by means of $R_{3}$.

Fig. 5. Practical circuit for three switched cut-off frequencies of 5,7 and $10 \mathrm{kc} / \mathrm{s}$. The insertion loss at low frequencies is approximately 6dB.

## NDVEMBER

## LONDON

9th. Television. Society.-"New techniques in receiver construction: printed circuits" by W. I. Flack at 7.0 at 164 Shaftesbury Avenue, W.C. 2.

13th. I.E.E. (Students).-"Digital computers and how they may help the engineer" by Dr. M. V. Wilkes at 6.30 at Savoy Place, W.C. 2.

14th. I.E.E.-" Frequency diversity in the reception of selectively fading binary frequency-modulated signals with special reference to long-distance radio-telegraphy" by J. W. Allnatt, E. D. J. Jones and H. B. Law. "An investigation of the spectra of binary frequency-modulated signals with various build-up waveforms" by J. W. Allnatt and E. D. J. Jones. "An improved fading machine" by H. B. Law, F. J. Lee, F. A. W. Levett and R. C. Looser. "The detectability of fading radiotelegraph signals in noise" by H. B. Law. "The signal/noise performance rating of receivers for longdistance synchronous radiotelegraph systems using frequency modulation" by H. B. Law. At 5.30 at Savoy Place, W.C. 2 .

14th. Radar Association.-"Infrared: its problems and possibilities" by Dr. F. E. Jones at 7.30 at the Anatomy Theatre, University College, Gower Street, W.C. 1 .

16th. B.S.R.A.-" Practical aspects of design and application of audio transformers" by R. B. Gilson at 7.15 at Royal Society of Arts, John Adam Sreet, W.C. 2 .

20th. I.E.E.-Discussion on "Data processing equipment for experimental work " at 5.30 at Savoy Place, W.C.2.

22nd. I.E.E.-Discussion on "The presentation and demonstration of the theory of semi-conductors to students " at 6.0 at Savoy Place, W.C.2.

22nd. Television Society.-"Alternatives to the N.T.S.C. colour system" by Dr. E. L. C. White at 7.0 at 164 Shaftesbury Avenue, W.C.2.
27th. Society of Instrument Tech-nology.-"Television technique applied to observation and control" by Professor J. D. McGee at 7.0 at Manson House, Portland Place, W.1.

28th. Brit. I.R.E.-"Colour Television" by Dr. G. N. Patchett at 6.30 at the School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.
30th. R.S.G.B.-Discussion on " $1250-\mathrm{Mc} / \mathrm{s}$ operation " at 6.30 at the I.E.E., Savoy Place, W.C. 2 .

## BIRMINGHAM

26th. I.E.E.-Informal evening on electronics and automation at 6.0 at the James Watt Memorial Institute, Great Charles Street.

## CAMBRIDGE

13th. I.E.E.-Address by Dr. R. C. G. Williams, Chairman, Radio and Telecommunication Section, at 8.0 at the Cavendish Laboratory.

20th. I.E.E.-Discussion on technical education to be opened by Professor E. B. Moullin and J. Wooding at 6.30 at the Cambridgeshire Technical College, Collier Road.

## CARDIFF

12th. I.E.E.-" Germanium and silicon power rectifiers" by T. H. Kin$\operatorname{man}$, G. A. Carrick, R. G. Hibberd and A. J. Blundell at 6.0 at the South Wales Institute of Engineers, Park Place.

## MEETINGS

## CATTERICK

22nd. I.E.E.-" Communication by tropospheric and ionospheric scatter" by Dr. J. A. Sazton and W. J. Bray at 6.15 at Catterick Camp.

## CHELTENHAM

21st. "Society of Instrument Techno-logy.-" Problems in the manufacture of semi-conductors" by F. C. Carpenter at 7.0 at the North Gloucestershire Technical College.

## EDINBURGH

6th. I.E.E.-" TRIDAC-a large analogue computing machine " by Lt.Cdr. F. R. J. Spearman, J. J. Gait, A. V. Hemingway and R. W. Hynes at 7.0 at the Carlton Hotel, North Bridge.

23rd. Brit. I.R.E.-"Information Theory" by L. C. Stenning, Dr. P. Jones and P. Holroyd at 7.0 at the Department of Natural Philosophy, University of Edinburgh.

## GLASGOW

7th. I.E.E.--"TRIDAC-a large analogue computing machine" at 7.0 at the Institution of Engineers and Shipbuilders, 39 Elmbank Crescent.

8th. Brit. I.R.E. -" The oscilloscope for engine testing" by R. K. Vinycomb at 7.0 at the Institution of Engineers and Shipbuilders, Elmbank Crescent.

## LIVERPOOL

7th. Brit. I.R.E.-" Industrial Television" by J. E. H. Brace and R. Swinden at 7.0 at the Chamber of Commerce, 1 Old Hall Street.

## LOUGHBOROUGF

20th. I.E.E.-" Ultrasonics in industry" by C. F. Brocklesby at 6.30 at Loughborough College.

## MALVERN

Ist. Brit. I.R.E.- "Principles of the light amplifier and allied devices" by Dr. T. B. Torilinson at 7.0 at the Winter Gardens.

## MANCHESTER

1st. Brit. I.R.E.-"Electronics applied to physiology " by H. W. Shipton at 6.30 at Reynolds Hall, College of Technology, Sackville Street.

7th. I.E.E.-Informal evening on electronics and automation at 6.45 at the Engineers' Club, Albert Square.

## NEWCASTLE-UPON-TYNE

14th. Brit. I.R.E.-"Some practical aspects of echo-sounding " by A. M. Sutton at 6.0 at Neville Hall, Westgate Road.

## OXFORD

14th. I.E.E.-"Automation and electronics in industry " by F. W. Highfield at 7.0 at 37 George Street.

## SHEFFIELD

21st. 1.E.E.-" Germanium and silicon power rectifiers" by T. H. Kinman, G. A. Carrick, R. G. Hibberd and A. J. Blundell at 6.30 at Grand Hotel.

## STONE

19th. I.E.E.-"The generation and synthesis of music by electrical means" by A. Douglas at 7.0 at Duncan Hall.

## WOLVERHAMPTON

14th. Brit. I.R.E.-" Electronic techniques in automation" by J. A. Sargrove
at 7.15 at the Wolverhampton and at 7.15 at the Wolverhampt
Staffordshire Technical College.


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# RANIDOM RADIATIIONS 

## By " DIALLIST"

## 819 Lines on Band I

THE first French 819-line television station operating in Band I was opened recently at Mont-Pinçon, near Caen. With its vision on $52.4 \mathrm{Mc} / \mathrm{s}$ and its sound on 41.25 $\mathrm{Mc} / \mathrm{s}$ the station needs a good deal of elbow room in the ether. It occupies actually a wider band of frequencies than our Channels 1 and 2 put together; but even so the $13-\mathrm{Mc} / \mathrm{s}$ modulation range normal for 819-line transmissions has had to be somewhat reduced which must, of course, affect the picture definition. The French broadcasting authorities were afraid that it might interfere in this country with Rowridge and adopted horizontal polarization as a safeguard against this. So far, I haven't heard of any such interference, even in this present period of freak radio effects; but Toute la Radio states that there has been some trouble to the south of Caen in places where Mont-Pinçon and Rowridge have about the same bearing.

## Slow to Catch On

THOUGH the development of the French television broadcasting network has gone forward quite rapidly, the increase in the number of TV receivers in that country has been curiously slow. There were roughly 220,000 sets in use a year ago and
the number now stands at not much over 350,000 . That's a 60 per cent increase, admittediy, but even so it looks as if much water will flow beneath the bridges of France before television becomes anything like the national hobby that it is in this country, where there are now well over $6,000,000$ receivers in use-the licence figure at the end of August was $6,044,330$.

## Nelson Effect?

IT takes some little time for $W$.W. to. make the journey to New Zealand that's why I have only just received from a kind Wellington reader two pieces of information which, throw a bright light on the "Bournemouth Effect." (June issue). That, if you remember, concerned an outburst of "chuffs" from a f.m. receiver every time a train left the station. One correspondent suggested that the effect might be due to a static charge generated by the friction of water particles driven at high velocity through air. The writer from New Zealand sends two proofs that this can happen. The first is a description published about 1870 by a lecturer at the Royal Polytechnic Institution of apparatus then used there for demonstrations and experiments. By means of a boiler, working at 601 b per square inch and provided with 46 bent iron tubes fitted


LLIFFE \& SONS LTD., Dorset House, Stamford Street, London, S.E. 1
with wooden nozzles, electricity was regularly generated and sparks up to 22 inches long were obtained. His second instance seems to clinch the point, for it shows the effects of such electric charges on wireless reception. At certain times every working day there was an outbreak of horrid noises from loudspeakers in Nelson, N.Z. It was a long time before an engineer who had been trying to find the cause stumbled on the solution. He happened one day to look out of the window whilst the noises were in full blast and saw a plume of steam issuing from a factory whistle some distance away. Subsequent observations confirmed that the outburst occurred only when the whistle was in action. No one had thought of the connection since owing to the distance travelled the sound of the whistle arrived after the noises had stopped. My correspondent suggests that since Nelson, N.Z., had the effect years before Bournemouth "Nelson Effect" would be a more appropriate term-provided that the Senior Service doesn't object!

## Canned Vision

A GOOD many attempts have been made to devise some workable system of recording vision on magnetic tape. Most, if not all, of these have proposed to record vision signals along a single straight-line path and one of the big snags encountered is that this would mean tape speeds of some 13,000 inches per second. In his recent application for a vision-on-tape recording system* Dr. R. D. A. Maurice, of the B.B.C. Engineering Division's Research Department, strikes out on an entirely new line. He proposes to scan a tape 1.3 inches wide by the use of what he terms "a magnetic Nipkow disc." This consists of a pair of copper discs, each containing 100 small equally-spaced slots filled with ferrite. The discs, whose distance apart is no more than about three-thousandths of an inch greater than the thickness of the tape, are locked together and driven by a motor at 6,000 r.p.m. This method would reduce the tape speed required to no more than 25 inches per second; but owing to the very

[^9]small aperture used the tape would have to be capable of retaining a maximum fux density fifty times that now used.

## Between Wind and Water

THE East-Coast town in which I now live has possibly had rather more than its fair share of the gales and gusty winds which have been amongst the more unpleasant features of 1956's apology for a summer. Very few aerials have been damaged by them, but a good many feeders which were not well enough anchored to prevent them from swaying to and fro have had to be replaced owing to breaks in their "inners." The most spectacular wind effect occurred here one afternoon when almost tropical rain was accompanied by gusts of the "Force 7 to 9 " variety. Each gust produced ear-splitting noises from loudspeakers and firework displays on television screens. Our electricity supply is conveyed to us by overhead three-phase mains on poles. At one place the wind was swaying the sodden branch of a tree on to the cables at frequent intervals.

## The Mystery that Wasn't

"THERE'S a television mystery in my house," said a friend the other day, "what happens is exactly the opposite to what you'd expect." When I asked for further enlightenment he told me that he'd had things so arranged that he could use his receiver in either of two rooms. "And here's the strange thing," he went on. "One of the rooms is much nearer to the aerial; but despite the much shorter length of feeder, you get nothing like so good a picture in it as you do in the other." I promised to drop in next time I was passing his way and when I did so I found exactly what I'd expected. The socket in the room nearer the aerial was simply Tee'd into the feeder. Hence with the set in use in that room there were yards of unused dead-end to upset the matching. The proper method is, of course, to connect the feeder to a socket in the nearer room and then to run a separate length to the other. At the far end of this length is a socket; the near end is fitted with a plug. If you want to use the set in the first room, you plug its aerial lead into the socket and the extra length of feeder is entirely unconnected. For reception in roem No. 2 you plug the eitension feeder into the first socket and the set's aerial lead into the second.


## UNBIASEID

## The 1957 Exhibition

ALTHOUGH it is almost a year before the next National Radio Exhibition opens its doors it is none to soon for the organizers to begin scratching their heads in an effort to think out how an improvement can be made on this year's effort.

Now without in the least decrying the efforts of the R.I.C. this year I can think of one very great improvement which I should like to see next August. To explain what I want, I should mention that whenever I wish to buy a set, tape recorder or what have you, I invariably go to the stand of one of the big wholesalers. The reason is that on these stands I can see the products of most of the manufacturers-some possibly, who are not themselves exhibiting at the Show. There I can compare sets side by side without rushing all over the exhibition.

It is, however, not easy for an ordinary member of the public to examine the goods on these stands. They are meant only for members of the industry and the gentlemen who staff some of them are apt to freeze off outsiders in the manner of a duchess dealing with a gatecrasher. I have no difficulty as I know enough of the trade jargon to be able to pass myself off as a dealer.

I would suggest that a large section of the hall be reserved for special stands each exhibiting one class of goods only. Each stand should be staffed by people who could be relied upon not to favour one manufacturer at the expense of others.

This idea would not prevent each manufacturer having a stand of his own as at present. These special stands would be extraordinary ones in the literal sense of that word. The idea is not new, of course. For some years we had a Television Avenue at each Show, where some dozens of sets could be seen operating, but, for some unknown reason, that disappeared from this year's show.
I hope the R.I.C. will consider my idea and if it does not receive favour, I hope the learned councillors will write and tell the Editor why.

## Literally Nostalgic

RECENTLY I had a personal letter from a friend living in the colonies deploring the fact that the sound of Big Ben has not been broadcast while the clock which is associated with this famous bell was out of action.
Apparently Great Tom proved an inadequate substitute. Dwellers in what used to be called the "far-flung Empire" have a sentimental interest in hearing the voice of Big Ben. It
has meant "home" to them for 30 years past and for once the B.B.C.'s overworked word "nostalgic" is literally correct.

It cannot be often that Big Ben has been out of action during the past three decades. Even if this is not the first time it is most certainly the longest period it has been silent. The B.B.C. had ample warning of the clock's long period of inaction and it seems a pity that they are so insensitive to the nostalgia associated with Big Ben in the colonies that they didn't bother to make a special recording of it. It would have been a simple technical matter for arrangements to have been made for the recording to have been triggered off at the appropriate times by signals from Greenwich.

The B.B.C. ought to be ashamed of themselves for not thinking of this idea instead of leaving me to do it for them,

## Wrinkles for ROSPA*

IN the September issue I discussed an article on road safety which appeared in the American journal Tele-Tech and Electronic Industries. Among other things it was suggested that a capacitative system should be used so that if a car approached too close to an obstacle ahead, the brakes would be applied. I concluded, however, the range of such a system would be far too small.

The ideal solution would be for cars to be fitted with radar as the range of operation would thereby be greatly increased. Unfortunately this would mean that the car would have to tow a trailer to house the necessary apparatus. I have actually been experimenting on these lines recently and my radar trailer attracts a good deal of attention as the high superstructure necessary to house all my experimental apparatus seems to the onlookers to block my rear view entirely and they gather with ghoulish glee at strategic road junctions in anticipation of a crash.

What the crowd don't know is that, in fact, I have a perfect $180-$ degree rear view. I have replaced the normal driving mirror in my car by a small TV screen which is coupled to a small camera in the rear of the trailer. I see that an American car at the London Motor Show is fitted with an "electronic rear viewer." Great minds think alike.

Talking of my radar trailer reminds me that one of the worst causes of car accidents is the slowmoving pantechnicon-type vehicle which completely blocks the forward

[^10]

A new rôle for TV.
view of following cars. The drivers cannot see the road ahead and dare not pull out to overtake. Eventually somebody loses patience and frequently a crash occurs. In my opinion this trouble could be overcome if all these juggernauts were compelled to carry a television camera in front, coupled to a screen in the rear, so giving drivers of following cars a clear view of the road ahead.

## Heard but Not Seen

WE are so used to hearing the clumsy, and ugly expression "loudspeaker" to describe what would be better called the reproducer or acoustic reproducer that I was considerably surprised recently to hear it spoken of as a "soft speaker." When I queried the name I was told that there was no mistake and this correctly describes the large number of instruments which are to be installed in Guildford cathedral.
This massive new building is slowly rising on a high hill outside Guildford. It is the last word in modernity and its p.a. system has been designed to be in keeping with it. The idea of the architect is that p.a. should be heard but not seen. Special pendant chandeliers are to be designed so that each will have what is to be termed a soft speaker built into it as an integral part of its structure. Each of these instruments will give a comparatively low acoustic output but, as there will be a large number of them, the overall volume will be adequate.

Needless to say, the designer is to collaborate with the leading lights in acoustics and in loudspeaker-or "soft speaker"-design.

# (10) <br> Regd. Trade Mark <br> WIDE BAND <br> SIGinal generator <br> Type T.F.M. 

The design of this new Wide Band Signal Generator, operating throughout on fundamentals, is the outcome of considerable research and development work to meet the stringent requirements imposed by new frequency modulation and commercial television stations.
 instrument not only meets all U.K. requirements, but will also prove useful in many other parts of the world.

## OUTPUT:

Minimum (about $2 \mu V$ ) to 100 mV conţinuously variable with decade multiplier. Force output 250 mV .

## OUTPUT IMPEDANCE:

$80 \Omega, 200 \Omega$, balanced $80 \Omega$ and $300 \Omega$, isolated unbalanced $80 \Omega$

The frequency bands have been chosen in such a manner as to ensure maximum convenience when servicing and aligning T.V. and F.M. receivers.

Provision has been made for spot R.F. frequency calibration.

Facilities are provided to ensure adequate discrimination throughout the very wide frequency band covered by the instrument.
Sine and square wave audio frequency modulation provided.

The instrument is fitted with an R.F. carrier level meter.

A double-ratio slow-motion mechanism, together with interpolation dial, enables the instrument to be set with a high degree of accuracy. On the F.M. range an internal phasing control enables the modulating signal to be applied to the X-plates of an oscillograph to produce a picture of a discriminator response curve.

## OPERATING VOLTAGES:

100-120, 200-260 V, $50-60 \mathrm{c} / \mathrm{s}$ A.C. mains.

DIMENSIONS:
$15 t \times 10 \frac{1}{2} \times 10 i n s$, approx, with lid closed.
WEIGHT: 16 lbs. approx. LIST PRICE : $\mathbf{t 8 9}$


In city streets, at speed on the open road, or crosscountry with the Army, transport communications must be completely dependable. As component manufacturers, we make an important contribution to that dependability.

The three MORGANITE Potentiometers illustrated are all completely sealed against moisture and dust.


Type LH.


Type H.
$1 \frac{1}{2}$ watts. For higher wattage requirements.


MORGANITE RESISTORS LIMITED
Bede Trading Estate, Jarrow, County Durham, England.

# RADAR DISPLAY <br> TUBES MITH Low Vottage Focus 

Easier setting-up Adjustment of electrostatic focus for absolute minimum of aberration can be made quickly and without special skill.

Simpler E.H.T. The focus voltage swing required is only $\pm 200 \mathrm{~V}$ about cathode-e.h.t. units need no longer be loaded with the current wasting potential dividers associated with earlier electrostatic focus tubes.

Lower cost No focus magnet or coil is required. Ordinary carbon track potentiometer across normal lowvoltage supply is all that is needed to achieve fine focus.

Space saved Focus at ordinary h.t. potential means that e.h.t. generators can be smaller, and bulky high voltage potential dividers eliminated.


| ABRIDGED |  | A T A | TUBES |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Typical Operating Conditions |  |  |  |
| Deflection: | Electrostatic Magnetic | $\mathrm{V}_{22+34}$ | $\mathrm{V}_{\text {al }}$ | $-V_{g} \text { for }$ | $V_{a 3}$ |
| Heater: | $6.3 \mathrm{~V}, 300 \mathrm{~mA}$ | (kV) | M | $(V)$ | $(\mathrm{V})$ |
| Phosphor: | long persistence | 12 | 300 | 30 to 70 | $\pm 200$ |

## Two New

## 4GP1

with a green screen having an afterglow of 100 milliseconds.

## 4GP11

with a photographic blue screen having an afterglow of 1 millisecond.

9 cm close tolerance, high quality instrument cathode ray tubes with electrostatic focus and deflection. These tubes have the following attractive features :-

1 Plate glass screen.
2 One stage of post deflection acceleration.

3 Low interelectrode capacitances.
4 Overcapped pressed glass wafer seal.
5 Orthogonality of deflection axes $\pm 1^{\circ}$ :
6 Spot centring. The undeflected spot will fall within a radius of 5 mms concentric with the tube face.

7 Deflection linearity. The plate sensitivity for a deflection of less than $75 \%$ of the useful scan will not differ from the plate sensitivity for a deflection of $25 \%$
of the useful scan by more than $2 \%$.

List Price 510.

> GE.C.
> Valves

## RATINGS

$V_{h} \quad 6.3 \mathrm{~V}$
in 0.5A.
$V_{a 4} \quad 8 \mathrm{kV}$ max.
$V_{a 3} \quad 4 \mathrm{kV}$ max.
$\left.\begin{array}{lll}S_{x} & \frac{620}{V_{a 3}} & \mathrm{~mm} / \mathrm{V} \\ S_{y} & \overline{400} & \mathrm{~mm} / \mathrm{V} 3\end{array}\right\} \begin{gathered}\text { When } \\ V_{24}=2 V_{a 3}\end{gathered}$

TYPICAL OPERATION
$V_{24} 4.0 \mathrm{kV}$
$V_{3} 3$ 2.0kV
$V_{a 2}$ (focus) 330 approx. $V$.
$\mathrm{Val}_{\mathrm{al}} 2.0 \mathrm{kV}$
$V_{g} \quad$ (for cut-off) -67 V
$S_{x} \quad 0.31 \mathrm{~mm} / \mathrm{V}$
$S_{y} \quad 0.2 \mathrm{~mm} / \mathrm{V}$
Line width 0.3 mms .

Additional technical information on these two new tubes may be obtained from the G.E.C. Valve and Electronics Dept.


IN great-grandma's time the old oak chest was generally the place where the household linen was stored. But to-day it has another use-at least in the home of Mr. Geoffrey Parnell, of Leeds. He writes: "Wondering how I could house the Ferrograph 66 I had on order, I suddenly noticed an old oak chest which had belonged to my wife's grandmother. It offered the ideal solution because not only did it take the Ferrograph but there was ample room, in addition, for a Garrard turntable and an F.M. Tuner unit".

Yes-you've certainly got something there, Mr. Parnell! The combination of these three units in one permanent installation-with an external loudspeaker-is the last word in home entertainment forecasting an important trend for the future. For not only does it permit tape recordings of the highest quality to be made from radio, disc or microphone at the touch of a switch, but the owner can use the high fidelity amplifier incorporated in the Ferrograph with the F.M. tuner as a superb radio receiver or with the turntable as a luxury gramophone. He has, in fact, three instruments in one at a very substantial saving in cost and space.

# FerNogmanh 



The remarkable efficiency of these 'Eclipse' magnets is due to their composite construction, using 'Araldite' to bond the component parts. The manufacturers of these magnets state that they use 'Araldite' because it enables them to produce shapes and sizes otherwise impracticable, to ensure that the magnets cannot be taken apart and to avoid bolted assemblies. 'Araldite' provides a bond which is truly permanent, and its strength is proved by the fact that facing and boring operations and also grinding are carried out after bonding.
'Araldite' epoxy resins have a remarkable range of characteristics and uses.

They are used

* for bonding metals, porcelain, glass etc.
$\star$ for casting high grade solid insulation.
$\star$ for impregnating, potting or sealing electrical windings and components.


## 'Araldite'

* for producing glass fibre laminates.
* for producing patterns, models, jigs and tools. $\star$ as fillers for sheet metal work.
$\star$ as protective coatings for metal, wood and ceramic surfaces.

'Araldite' is a registered trade name


## Distortion detectedTransmission unaffected

## with the T. D. M.S.

The T.D.M.S. 5A and 6A are portable sets designed to measure. distortion at any point in a radio teleprinter or line telegraph circuit without interfering with normal transmission. The equipment consists of two units each $18 \frac{1}{2}^{\prime \prime} \times 11 \frac{1}{2}^{\prime \prime} \times 13 \frac{1}{2}^{\prime \prime}$ both mains driven and electronically controlled. Either may be used independently for certain tests or both may be used in combination to cover a comprehensive range of testing operations.

T.D.A.S. 5A

Sends an automatic test message, or characters, or reversals at any speed between 20-80 bauds with or without distortion. The CRO has a circular time base for distortion measurements on synchronous signals only, or relay adjustment. Weight 37 lb .

You are invited to apply for a copy of a descriptive leaflet.

## AUTOMATIC TELEPHONE \& ELECTRIG CO. LTD.,

RADIO AND TRANSMISSION DIVISION,
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# ARHE COMMUNICATION RECEIVER Type C864 

## It's NEW-A RECEIVER with an astonishing PRRFORMANCE/PRICE RATIO

ALTHOUGH priced at only $£ 120$ the AIRMEC COMMUNICATION RECEIVER TYPE C864 has a specification equal to that of many sets selling at double its price. Some of the main features are:-

- Frequency coverage from $15-45 \mathrm{kc} / \mathrm{s}$ and $100 \mathrm{kc} / \mathrm{s}-30 \mathrm{Mc} / \mathrm{s}$.
- Film Scale giving actual Scale length of 4 ft . on each frequency range.
- 90 : I Slow Motion Drive with logging scale.
- Crystal Calibrator incorporated.
- Frequency setting accuracy better than $1 \mathrm{kc} / \mathrm{s}$.
- Separate Incremental tuning control for use with Crystal Calibrator.
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- S Meter incoroporated.
- Very stable B.F.O.
- Muting facilities provided
- Built-in Loudspeaker.
- 2 Watts Output.
- Turret band switching.

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Write now for full details to :


## Thy Appointment to the Diofessional Enginees 'MULTICON' PLUGS AND SOCKETS


range consists of $2,4,6,8,10,12,18,24$ and 33 -pole sizes, and there is a complete variety of Plugs and Sockets with alternative mounting arrangements.
cable-fixings, retaining devices and cover earthing facilities.
ELECTRICAL SPECIFICATION AND PERFORMANCE

## VOLTAGE RATING <br> (a) 1,000 Volts D.C. or A.C. (peak).

This applies to use in temperate climates under normal conditions.
(b) 500 Volts D.C. or A.C. (peak).

This applies to tropical use.
2. Voltage Proof:

All plugs and sockets will withstand a voltage proof test of 2.5 KV between contacts, and 3.0 KV between contacts and the mounting flange or cover.
3. Breakdown or Flashover Voltage :

The breakdown voltage is approximately 3.3 KV .
The properties of the mouldings are such that even after flashover there is no evidence of tracking, and the plugs and sockets will still withstand I the Voltage Proof specified above.
CURRENT RATING CONTACT RESISTANCE

5 Amps. D.C. or A.C. (R.M.S.) per contact. 1. Average Contact Resistance: Below 0.002 Ohm. 1 2. Maximum Contact Resistance : 0.0025 Ohm .
(1) DESIGN

The terminal numbering is moulded into both the plug and socket bodies, and appears not only in proximity to the appropriate soldering-tag. but also on the mating face.
This not only facilitates wiring, but enables complete cable forms to be tested prior to inclusion in equipments without removing the plug or socket covers.
(2)

Four smali distance pips are moulded on to the plug body and they keep the mating faces slightly apart even when the plug's and sockets are fully engaged.
This eliminates the possibility of free molsture remoining between the plug and socket foce, and is instrumentol the superior tropical performance of "Multicon "plugs and sockets.

## (3)

The single-piece, body mouldings are nylon-filled to provide a high insulotion and trocking resistdnce.

## (4)

All socket clips and plug blades are located in recessed cavities in the mouldings.
This also provides a high tracking resistance between contocts, and, in the socket version, the enclosed contacts enable the maximum voltage to be safely utilised (provided the direction of voltage supply feed is from socket to plug).

## (5)

Each socket clip has split limbs, so that there are four indlvidual areas in contact with each plug blade.
This ensures obsolute reliabllity of contoct, with a minimum life of 10,000 operations at low ond constont contact resistonce.

FEATURES
(6) Panel or chassis mounting flanges are available with side or end fixing holes. Cut-out information given relates to Panel thicknesses up to $\frac{3^{3}}{32^{\prime \prime}}$.
This? enables rows of plugs or sockets to be mounted either end-to-end or side-by-side with a maximum saving in panel or chassis space.
(7)

In the 24 and 33 -pole sizes the plugs are normaily supplied with a large pin locator, which is also an electrical contact.
This is provided to foclitote the engagement of these larger sizes, especiolly in unitor opplications.
(8)

Covers are provided with either a top cable-entry hole and clamp or a side cable-entry hole and clamp, to suit the needs of porticular equipments and designs.
(9)

Two alternative facilities can be provided for earthing the plug or socket (9A.).
In one version an earthing tag attached to the moulding connects the inside of to the moulding connects the inside of
the cover to the highest numbered conthe cover to the highest numbered con-
tact so that an earth lead in the cableform tact so that an earth lead in the cableform,
connected to the highest numbered contact, automatically earths the cover. (9B.)
In the alternative form, an earth tag is rivetted directly to the outside of the cover and is suitable for the direct connection of an earth lead.
By either methad the cover is corthed to ensure the sofety of the users of equipments in which "Multicon" plugs and sockets are incorporated.

## (10)

All sizes of plug or socket covers can be fitted with retaining blades to secure the unit to the panel or chassis.
Even under the most severe vibration conditions. therefore, or in the cose of accidental interference with the cobleform, there is absolute reliabillty of contact.

(I)

(2)

(7)




## 3 amplitude modulated versions are available

$\mathrm{W}=100 \mathrm{Kc} / \mathrm{s}$. channelling for aeronautical and multicarrier schemes.
$\mathrm{N}=60 \mathrm{Kc} / \mathrm{s}$., $50 \mathrm{Kc} / \mathrm{s}$. or $40 \mathrm{Kc} / \mathrm{s}$. channelling. $\mathrm{VN}=30 \mathrm{Kc} / \mathrm{s} ., 25 \mathrm{Kc} / \mathrm{s}$. or $20 \mathrm{Kc} / \mathrm{s}$. channelling.

The Ranger has been defsigned to meet the following leading specifications.
U.S. Federal Communications Commission Canadian R.E.T.M.A.
Canadian Dept. of Transport
British G.P.O. existing and proposed specifications

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Pye Limited, Mexico City.
Pye Limited, Tucuman 829, Buenos Alres, Argentina.

Pye Radio \& Television (Pty.) Ltd. Johannesburg, South Africa.
Pye (Canada) Ltd., 82, Northline Road, Toronto, Canada.
Pye Pty. Ltd., Melbourne, Australia.

Messrs. Telecommunications, Jamestown Road, Finglas, Co. Dublin.

Pye Ltd.
Auckland, C.I., New Zealand,
Pye (Frarice) S.A.
29 Rue Cambon, Paris Ier


## conclusive evidence

Tinker, tailor, soldier, sailor (or for that matter, musicians, airmen, teachers, chemists) sitting in judgment on the latest high fidelity equipment in our showrooms, where $\mathrm{Hi}-\mathrm{Fi}$ is in the witness box every day telling the whole truth and nothing but the truth.

All the evidence you need to reach a verdict is here at your disposal.

Be your own judge and jury.
Come along to Imhofs and listen to the evidence of your own two ears.

Write or phone for a copy of our new Hi-Fi Catalogue
or better still call in and pick one up. It is.
comprehensive, fully illustrated and completely free


Power Controls Ltd., Exning Read, Newmarket, Sutiolk
Tolephone: Nowmarket 3181. Telegrams: Powercon, Nowmarkot

Have you a transformer problem? If so, we can help you. We can undertake to develop and manufacture rotary transformers to your specification. The illustration shows a typical transformer which we are manufacturing for a specific requirement. Made for 6,12 or 24 volts D.C. input, it can supply a continuous D.C. output of 350 volts at 30 mA . or an intermittent output of 310 volts at 60 mA . The no-load current consumption is 2.2 amps . at 11.5 volts and the ripple voltage is less than 6 volts r.m.s. on 60 mA . load. The size is only $4-9 / 16^{\prime \prime}$ long by $2-21 / 32^{\prime \prime}$ across the brush terminals.


You need to be able to live with your high fidelity equipment as well as listen to it.
This RCA New Orthophonic High Fidelity speaker system uses advanced acoustical design to give a liveable size with clean aesthetic lines which will harmonise with any furnishing scheme. The system uses a ported bass reflex type enclosure including the exclusive RCA acoustic curtain damping feature. The special arrangement of the triple speaker assembly ensures even sound distribution to all parts of your room with perfectly balanced bass and treble response. This instrument with its great sensitivity captures those elusive elements of sound which make for completely natural reproduction. The RCA Loudspeaker is the latest addition to a complete range of perfectly matched High Fidelity units designed for studio quality, radio and gramophone reproduction in the home. Send for further details.

## Their CONSTANT PERFORMANCE....


H. F. HEATING


BEAM TRANSMISSION


SERVO CONTROL



INSTRUMENTATION



Standard models are available covering power requirements from 4 watts to 6 kilowatts.

IIn much of today's highly sensitive electronic equipment and processes consistent high performance is possible only if the supply voltage is stabilised to the specific value for which the equipment is designed. Without constant voltage, instruments become unreliable, communications are interrupted, process control becomes erratic, products become degraded and discredit falls on the manufacturer. Advance Constant Voltage Transformers provide the answer to these problems. Entirely automatic, with no moving parts, they can accommodate wide voltage variations in the supply over a range of $\pm 15 \%$ of nominal, providing a stabilised voltage within $\pm 1 \%$.

AC voltage stabilisation is our business and our technical representatives are at your service if you encounter difficulties of this nature. We shall be most pleased to forward to you fullest details upon request. Write for folder W.28.


Type GB 896
(A.R.B. Ref. No. E3518) For loads from 2 to 35 lbs
"BARRYMOUNT" Air-damped Isolators have been specially developed to provide assured protection for sensitive equipment against vibration and shock. An outstanding feature is their remarkably uniform performance over the full rated load-range.
Over $1,000,000$ go into use every year for the protection of every type of air-borne equipment, from the lightest and most delicate instruments and electronic devices to apparatus up to I 40 lbs. weight.

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- Improved shock absorption.
- Low permanent set and drift.
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Wide load range with uniform performance.


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Ahead of the present -



## INSTRUMENTS

## ADMITTANCE BRIDGE TYPE B. 801

$$
1 \mathrm{Mc} / \mathrm{s}-100 \mathrm{Mc} / \mathrm{s}
$$

For balanced and unbalanced measurement.
Susceptance: Equivalent
Conductance : 0-100 mmho. to $\pm 230 \mathrm{pF}$.
Accuracy: $\pm 2 \%, \pm 0.5 \mathrm{pF}$.
Accuracy: $\pm 2 \%, \pm 0.1 \mathrm{mmho}$. This is one of a range of bridges for use with external source and detector for the measurement of aerials, cables, feeders, and a variety of components and materials between $15 \mathrm{kc} / \mathrm{s}$ and $250 \mathrm{Mc} / \mathrm{s}$. Bridge sources and detectors are available for use between $1-100 \mathrm{Mc} / \mathrm{s}$ and $50-250 \mathrm{Mc} / \mathrm{s}$.

PRICE 1150 NET EX WORKS


## AUDIO WAVEFORM ANALYSER TYPE A. 321

A portable instrument to measure the relative levels of the components of a complex waveform over a range of 75 db between $50 \mathrm{c} / \mathrm{s}$ and $20 \mathrm{kc} / \mathrm{s}$. Input impedance $100 \mathrm{~K} \Omega$ unbalanced or $>25 \mathrm{~K} \Omega$ balanced. In transportable case as shown, or for standard $19^{\prime \prime}$ mounting.

PRICE £250 NET EX WORKS

COMPONENT BRIDGE TYPE B. 121
A general purpose 50 cps 3 terminal transformer ratio arm bridge for the measurement of Resistance, Capacitance and Inductance in the ranges $3-100 \mathrm{M} \Omega$, $1 \mathrm{pF}-100 \mu \mathrm{~F}$ and $100 \mathrm{mH}-10,000 \mathrm{H}$, accuracy $\pm 2 \%$. Direct readings of the resistive and reactive components of impedance and facilities for "in situ" measurements are notable features.

PRICE £60 NET EX WORKS

## Sentercel HT. RBCTITITBRS from 125 V 30 mA

Specially designed for use in domestic Radio \& Television receivers, these miniature rectifier stacks have an established position with manufacturers to whom reliability, small dimensions and low costs are important.

| TYPE | RMO | RMI | RM2 | RM3 | RM4 | *RMS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Maximum amblent temperature <br> Maxlmum output current (mean) <br> Maximum input voltage (r.m.s.) <br> Maximum peak lnverse voltage <br> Max. Instantaneous peak current <br> Weight | $35^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C}$ <br> 30 mA 15 mA 125 V 350 V Unlimited 0.820 oz | $35^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C}$ 60 mA 30 mA 125 V 350 V Unllmited 1 oz. | $\begin{array}{\|c} 35^{\circ} \mathrm{C} \\ 100 \mathrm{~mA} \\ 10{ }^{\circ} \mathrm{C} \\ 125 \mathrm{~mA} \\ 350 \mathrm{~V} \\ \text { Unlimited } \\ 1.4 \mathrm{oz} \end{array}$ | $\begin{array}{cc} 35^{\circ} \mathrm{C} & 55^{\circ} \mathrm{C} \\ 120 \mathrm{~mA} & 90 \mathrm{~mA} \\ 125 \mathrm{~V} \\ 350 \mathrm{C} \\ \text { Unfilited } \\ 202 . \end{array}$ | $40^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C}$ 250 mA 125 mA 250 V 700 V Unlimited 4.5 oz. | $\begin{gathered} 40^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C} \\ 300 \mathrm{~mA} \quad 150 \mathrm{~mA} \\ 250 \mathrm{~V} \\ 700 \mathrm{~V} \\ \text { Unllmited } \\ 4.75 \mathrm{oz} \end{gathered}$ |

- For use in voltage-doubler circults the peak Inverse and maximum input voltages are halved, current outpus being as for half wave operation.
- Instant starting-no warming-up period
- Unlimited instantaneous overload
- No limit to size of reservoir capacitor
- Simple mounting-no valve holder
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- Practically indestructible in service
- Simple wiring-fwo connectors only

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## RTDIO EXPORT

## TUBES ONLY

1900 types of both receiving and transmitting tubes in stock. In addition, a comprehensive range of crystals and some types of transistors and trustworthy tubes are available.
Our new price list shows considerable price reductions on many current types. To these have been added many new types.

PRICE AND STOCK LISTS MAY BE HAD ON APPLICATION.
Your specific enquiries for special types to CV., JAN and MIL specifications are invited.

Our organisation is A.R.B. approved.


$\star$ The 595 HS can be controlled by ultra sensitive contacts handling 0.4 mA . at 2 V . Contacts will handle 5 A, at 230 V. A.C. $\star$ The 595 HS is made to withstand exceptionally heavy shock and vibration. * The 595 HS is made to withstand dirt and humidity indefinitely.

* The 595 HS can be obtained with various contact assemblies.
$\star$ The 595 HS is low in price because of its novel design.


## for the closest approach to the original sound



The criterion, as always, is that the reproduced sound shall be the closest
approach to the original - that the enjoyment and appreciation of music may be unimpeded. This is reflected throughout the design of the QUAD II. It is reflected,
too, in the straightforward and logical system of control,
achieved without the sacrifice of a single refinement or adjustment capable of contributing to the inal objective.

Send for further details and booklet:


## VACANCIES FOR ENGINEERS

If you have a genuine interest in Audio Engineering and a sound technical background, we will be pleased to discuss the possibility of your joining our Company.

## WIDE RANGE CAPACITANCE BRIDGE



Designed for the accurate measurement of capacitance and resistance in the range -0.002 pF to $100 \mu \mathrm{~F}$ and $/ \Omega$ to $10,000 \mathrm{M} \Omega$ respectively.
All measurements are made in the form of a three terminal network and components can be measured in situ. Accuracy within $\pm 1 \%$ Frequency $1592 \mathrm{c} / \mathrm{s}(\omega=10,000)$.
Full technical information on this and other 'Cintel' Bridges is available on request.

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## PARMIEKO TRANSDUCTORS for MAGNETIC AMPLIFIERS



## television monitor comprehensive

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- Metal backed screen.
- Straight gun (no ion trap) permitting highest spot quality.
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- Screen blemishes reduced to the absolute minimum.
- Quality control throughout manufacture to ensure highest picture standards and maximum life.
- Magnetic and electrostatic focused equivalent available to suit individual design requirements.


AW13-36 Electrostatic Focus
Max. final anode voltage 14 kV Typical focus voltage range -200 V to +200 V about cathode potential
Deflection angle $53^{\circ}$
Overall length 12 inches


MW13-35 Magnetic Focus
Maximum final anode voltage 11 kV
Deflection angle $53^{\circ}$
Overall length 11 inches

## ( 5 -inch TUBES

These tubes are designed for use as electronic view-finders in television cameras, but they also satisfy the requirements for compact monitor equipment in broadcast and industrial television.

## 14-inch TUBES

These tubes are designed for use in television studio monitors but they also satisfy the requirements for large screen displays in industrial television systems.

# Mullard 

COMMUNICATIONS AND INDUSTRIAL VALVE DEPARTMENT


AW22-10 Electrostatic Focus

Max. final anode voltage 14 kV
Typical focus voltage range -200 V to +200 V about cathode potential
Deflection angle $58^{\circ}$
Overall length 16 inches

## 17-inch TUBE

Max. final anode voltage 14 kV Deflection angle $64^{\circ}$
Overall length 15 inches

## ( 9-inch TUBES

These tubes are designed to a size convenient for use in mobile outside broadcast television equipment. They are also employed in studio floor monitors.

MW22-22
Magnetic Focus


MW36-67
Magnetic Focus
Max. final anode voltage 15 kV Deflection angle $70^{\circ}$
Overall length $17 \frac{1}{2}$ inches


This tube is primarily intended for use as a television studio monitor tube.

## MW43-67 <br> Magnetic Focus

Max. final anode voltage $15 \mathbf{k} V$
Deflection angle $70^{\circ}$
Overall length 20 inches


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* Combined conduction and convection cooling


## Longer life through improved cooling

The problem of heat dissipation has be'en neatly solved.
Double cooling* increases reliability and lengthens life.
A television component of outstanding dependabilityl

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An extremely wide range of Dubilier resistors is available both for development and production purposes. This range covers insulated wire-wound, power wire-wound, precision wire-wound, ultra high range, high stability, high voltage and high frequency resistors.

DOBIMTB8

## Hi-Fi Fans

 besiege Stentorian Dem. Room at Radio Show rom morning to night the Stentorian Demonstration Room-with a limited seating accommodation for only 50 people at a time-was one of the main attractions at Earls Court. For word had got around that W.B. was demonstrating Hi-Fi of outstanding quality at prices that were astonishingly low.And for once Dame Rumour was no lying jade! During the period of the Exhibition, in fact, no fewer than 11,000 Hi-Fi fans heard for themselves - and were convinced - that there was no finer equipment at any price than Stentorian.

Time and again people came back and said : "We've heard other demonstrations but yours is the best". Of course, we are gratified at the enthusiasm shown, but not surprised. For we've been in the Loud Speaker business right from the beginning. Everything that goes into a W.B. Speaker has been designed and made in our Mansfield works. We even make our own magnets.

We maintain a fine team of research engineers with many 'firsts' to their credit. Perhaps you did not


ENTRANCE
1 Stentorian H.F. 816 in Junior Bass Reflex Corner Cabinet
2 Stentorian H.F. 1012 in standard Bass Reflex Console Cabinet 3 W.B. $12^{\prime \prime}$ Concentric Duplex in special Cabinet
4 Stentorian H.F. 1214 standard Bass Reflex Cabinet with T. 816 .

5 Stentorian H.F. 1012 with T. 10 tweeter in Senior Bass Reflex Corner Cabinet
6 Stentorian H.F. 812 in Corner Console Cabinet
The demonstrations were carried out in conjunction with the W.B. 12 Amplifier and VHF/FM Tuner which were housed, together with a record player, in a W.B. Hi-Fi Console Cabinet.
know that the first "tweeter" ever to be produced in this country was made by us. But if W.B. has always been associated with high quality-and real Hi-Fi can only be achieved with high quality equipment-we have always set our face against high prices.

Compare the prices shown here and you'll appreciate that Hi-Fi is now available at realistic cost. See and hear Stentorian Hi-Fi at your Dealer's - we are content to await your verdict.

Type H.F. 1012
$10^{\prime \prime}$ unit, dle cast, 12,000 gauss magnet, cambric cone, 10 watts capacity. $30-14,000$ c.p.s. Bass


Type H.F. 812
$8^{\prime \prime}$ unit, 12,000 gauss magnet, cambric cone, 5 wates capacity, 50-12,000 c.p.s. Bass resonance 65 c.p.s. Die cast chassis:
44.3 .6

## Type H.F. 816

$8^{\prime \prime}$ unit, die cast, 16,000 gauss magnet, cambric cone, 6 wates capacity. $50-14,000$ c.p.s. Bass resonance 63 c.p.s. 66.17 .0

Type T. 816
Special $8^{\prime \prime}$ mid-range unit for use with H.F. 1214, 16,000 gauss magnet, 15 watts capacity with 1,500 c.p.s. crossoover. Up to 17,000 c.p.s. Impedance i5 ohms. C6. 10.0


Type H.F. 1214
$12^{*}$ unit, die cast, 14,000 gauss magnet, cambrle cone, 15 watts capacity. 25-14,000 c.p.s. Bass resonance 39 c.p.s. 69. 15.6

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Tweeter init, m/c pressure type, 14,000 gauss magnet, 2,000-14.000 c.p.s. 5 watts. Recommended for use with H.F. 1012. 44.4 .0

12* Concentric Duplex
Combined bass and weeter unit with embric cone and mid-range frequency stabilizers. Handling capacity, 15 watts. Frequency
res. 25 c.p.s. to 17,000 response, 25 c.p.s. 20 . 37 c.00 ce, 25 c.p.s.
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Stentorian W.B. 12
Quality Amplifier 12 watts, low nolse input clrcult, double trlode phase splitter, push-pull output stage glving outstanding re-
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Corner Console
$26^{\prime \prime} \times 17^{\prime \prime} \times 7 \frac{1}{4}^{*}$ for use with Stentorlan $H_{0}$ F, 812 E5. 10.0


Junior Bass Reflex Corner Console
For use with $8^{\circ}$ or $10^{\circ}$ units with provision for tweeter.
$33^{\prime \prime} \times 221^{\prime \prime} \times 18$ 月 $^{*}$
$£ 9.9 .0$


Senior Bass Reflex Corner Console
For use with $10^{\circ}$ or $12^{\circ}$ units with provision for eweeter. $35^{*} \times 30^{n} \times 19^{*} 611.11 .0$


Standard Bass Reflex

## Console

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$32^{\prime \prime} \times 22^{\prime \prime} \times 16^{\prime \prime}$. 40.10 .0

Al prices in this. advertisement are inclusive of Purchase Tax.


See and hear these and all other W.B. Ilnes at our London office ( 109 KIngsway , WC2) any Saturday between 9 a.m. and 12 noon. Leaflets of all the outstanding W.B. products on request.



SCALE S2

SCALE S2. For use with 500 pF . Tuning Condenser.
This Glass Scale is printed in Yellow with Long, Medium and Short Wavebands and a 0-100 Logging Scale. Station names, Amateur and Broadcast Bands are prominently marked. Designed for use with Coil Packs CP.3/500, CP.3/G, CP.3/F, CP.3F/G and also 500 pF tuning coils. Very suitable for use with a 3 Waveband Coil Pack (CP.3/500 or CP.3/G) leaving the Log Scale for tuning a V.H.F./F.M. Tuner.

Scale coverage: Long Wave $800-2,000$ metres.

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\begin{array}{lc}
\text { Med. Wave } & 200-550 \text { metres. } \\
\text { Short Wave } & 16-50 \text { metres. } \\
\text { Log } & 0-100 .
\end{array}
$$

The Scale measures $8 \frac{1}{8} \mathrm{in}$. $\times 6 \frac{7}{8} \mathrm{in}$., and is for a cabinet aperture of $6 \frac{1}{2} \mathrm{in}$. $\times 5 \frac{5}{8} \mathrm{in}$.
The Kit comprises of: Glass Scale, Back Plate, Pulleys Rubber Scale Mounts, Pointer, Drive Cord, 4 B.A. Screws Nuts, Spacers and Assembly Instructions.

PRICE 15/-
SCALE S1. "MAXI-Q "Basic 5 Waveband Glass Scale as above for use with 315 pF Tuning Condenser.
This five-colour glass scale covers the following bands: Long Wave $150-400 \mathrm{Kc} / \mathrm{s}$, Green; Medium Wave 530 1,600 Kc/s, Red; SW. $11.5-4 \mathrm{Mc} / \mathrm{s}$, White; SW. $24-13 \mathrm{Mc} / \mathrm{s}$, Blue; SW. $310-30 \mathrm{Mc} / \mathrm{s}$, Yellow.

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12 2 in .
$6 \frac{1}{2} \mathrm{in}$.

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In design, dependability, accuracy and freedom from wear these Egen components are quite outstanding. They are backed by unrivalled experience of the requirements of television and electronic equipment manufacturers.


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with concentric operating spindles. Incorporating many outstanding design features, including multiple contact rotors and thorough screening between sections. Control spindles can be supplied to suit customers' requirements. Type 136 less switch. Type 137 with SPST switch. Type 138 with DPST switch.

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Completely enclosed in bigh-grade phenolic mouldings. Solder tags heavily silver-plated for quick soldering. Fully insulated spindles with integral control knobs. Tapped for 2 -hole 6 B.A. fixing on ${ }^{\frac{1}{p} \text { " centres. Type 126, }}$ wire-wound. Type 127, carbon.


## MINIATURE POTENTIOMETERS

${ }^{\frac{7}{8} / \prime}$ diameter: utmost reliability within a very small compass. Positively located soldering tags, silverplated for easy soldering. All steel parts rustproof. Standard values available, from 5000 ohms to 2 megohms. Type 115 less switch, Type 105 with specially designed 2-pole Q.M.B. switch.


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Of robust construction; its unique design of laminations enables Transformer characteristics to be closely controlled, despite its small size. Weight: $\cdot \mathbf{1 2}$ ozs. Lead-out wires are colour coded for phase sense of windings applications.
Specially designed apparatus, capable of detecting one shortcircuited turn of 50 SWG in 5,000 turns, is used in manufacture to safeguard against premature failure caused by electrolytic action on the fine wire at short-circuited points.
A comprehensive range is available. A typical example is a Transformer with primary inductance of 100 Henries with no D.C. in the winding. The graph shows variation of primary inductance with D.C. in a similar unit.
Special designs of Transformers are produced to meet individual requirements.

## THE MINIATURE EARPHONE

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## Outstanding range <br> of Thorn <br> miniature



## miniature sealed panel lampholder - indicator type

Completely waterproof and will withstand conditions of constant vibration and shock, these lampholders are intended for installation on aircraft, armoured fighting vehicles, and marine equipment. They are sealed and insulated from the panel, the thickness of which can vary from 20 S.W.G. (.036") to 10 S.W.G. (.128\%). Thicker panels can be counterbored. Rotation is prevented by flats on the body. Mounting is by a single hole. Access to the lamp, for replacement, is from the front of the unit by unscrewing the dome. Lamps may be renewed, without breaking the seal to the equipment.
Weight : . 420 oz . ( 11.6 grammes) with bulb.
Electrical connections: Two solder tags.
Catalogue No. MPL. 20 Red : MPL. 21 Green.
Catalogue No. MPL. 22 Amber : MPL. 23 Opalescent Ivory.

## MINIATURE SEALED PANEL LAMPHOLDER - DIMMER TYPE

Identical to the Indicator type, except for the interchangeable cap. This is ribbed for grip, continuously rotatable and contains a light output control from bright to 'blackout', Weight : 530 oz . ( 14.8 grammes) with bulb. Electrical connections : Two solder tags. Catalogue No. MPL. 10 Red (Translucent). Catalogue No. MPL. 11 Green (Transparent). Catalogue No. MPL. 12 Amber (Transparent).
Catalogue No. MPL. 13 Clear (Transparent).

## THORN MIDGET PANEL LAMPHOLDER

This is the simplest and most economical lampholder designed to accommodate the Atlas Midget Panel lamp. It is extremely effective and easily installed. Available with its transparent top in a variety of colours. Weight : 8.4 gr . ( 0.3 ozs .)
Can be supplied with insulated washers and connecting tags where non-earth return is desirable.

## Miniature lampholders in the Thorn range

have been made possible by the development of the Atlas Midget Panel bulb.
1.5 volts 0.75 amps


## Thome

Aircraft Components Division, Gt. Cambridge Road, Enfield, Middlesex.

## FLUSH OR RECESSED LIGHTING UNIT

This lampholder is used as a standard unit in the Plasteck Console panel. The body of the lampholder may be retained in a countersunk hole in the panel by a hexagonal backnut and lock-washer. A small projection under the collar prevents the fitting turning in the panel. The special coloured filter is contained in a moulded screw cap and a soft rubber sealing washer prevents any light from escaping round the edge. Filters in red, green, amber and clear. Weight: . 31 oz . with bulb.
Terminals: Solder tag and earth return.
Catalogue No. PPL90.
Catalogue No. PPL120 (with 6BA terminal screw and earth return, weight : .35 oz . with bulb).
Interservice ref: Type A, No. 1.
Flush type - Solder connections. Ref. No. 5C/X. 5143.
Type A, No. 2.
Flush type-Screw terminals. Ref. No. 5C/X. 5144.
Can be supplied with insulated washers and connecting tags where non-earth return is desirable.

## SURFACE TYPE LIGHTING UNIT

An alternative design to PPL90 for Plasteck and other control panels where no room exists immediately behind the metal panel. The bulk of the component projects above the face of the panel. A soft rubber sealing washer under the cap prevents the escape of light from the front of the panel. The lamp is inserted with the cap up.
Weight: .49 oz . with bulb.
Terminals: Solder tag and earth return.
Catalogue No. PPL. 100.
Interservice Ref : Type B,
Surface type-Ref. No. 5C/X. 5145.

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## Have the following to offer from stock:



This has 9 valves, 6 -watts push-pull output, piano key selectors, independent bass and treble controls and magic-eye tuner. 28 gns . cash or $£ 6 / 8 /$ - deposit and 8 monthly payments of $63 / 9$.

FOR PICK-UP PERFECTION
The New B. J.
Super 90. 2 plug-
in shells supplied
with each arm,
popular cartridges, $£ 16 / 6 / 5$.
Pye HF25/SC HI/F1, Loudspeaker system consists of a high grade dual concentric loudspeaker housed in an elegant corner cabinet. Frequency coverage 30 to over $20,000 \mathrm{cps}$., 25 w . undistorted output. 65 gns.

> Part exchange is our speciality. Write giving details of your present equipment, and we will give you a quotation by return.

Jason MW/FM tuner with A.F.C. A highly sensitive self-powered unit with $500 \mathrm{~m} . \mathrm{V}$. output, $£ 28 / 13 / 6$ cash or $\mathbf{£} 6 / 13 / 6$ deposit and 8 monthly payments of $61 / 3$.

Lowther LLIO linear amplifier complete with control unit, $£ \mathbf{5 7} / \mathrm{lo} / \mathrm{-}$. Vortexion tape recorder with Wearite deck, $\mathbf{6} 84$ cash or $\mathbf{6 4 2}$ deposit and 18 monthly payments of 53/8.
T.S.L. Lorenz 12 in . Diaxial Speaker combined with two L.P.H', 65 high frequency treble units mounted coaxially to give a frequency range of 20-22,500 c.p.s, $£ 14 / 19 / 6$.

## SPECIAL OFFERS

Accoustical Quad II Amplifier and control unit, as new, $£ 35$. Accoustical Quad I Amplifier and control unit, $£ 18$.
Lower PWI Corner Horn complete with PMI pressure unit. Walnut veneered cabinet, $\mathbf{E 3 5}$.
Wharfedale $W$ is CS Speaker, $£ 13$.
Wharfedale Super 8 CSAL, $£ 6$.
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[^11]
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Incorporating a germanium
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Encapsulated in a seamless metal can, its dimensions are only- $4.75 \times 3.1 \times 7.5 \mathrm{~mm}$.

## A brilliant new range Packaged Hi Fi made by CRAFTSMEN



With small, matching and attractively presented plug-in units, Pye is bringing high fidelity to the ordinary listener. This is being done with absolutely no reduction in the quality which has already won acclaim for Pye High Fidelity Systems in over fifty countries. A modern Hi Fi amplifier requires a high degree of manufacturing skill to ensure that it reaches its exacting specification. A Pye Group Company in London specialises in amplifier production and employs carefully selected craftsmen who are expert at complex wiring and assembly. This is your guarantee that all Pye equipment leaving the factory is of the very highest standard.

## HIGGE FIDELITY STSTEMS

The New Sound in Home Entertainment

Pye Limited, Auckland, New Zealand;

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Pye Corporation of America, 270 Park Avanue, New York;

Pye LImited, Tucuman 829, Buenos Alras: Pye Limited, Mexico City;

Pye Radio \& Television (Pty.), Limited, Johannesburg, South Africa;

Pye LImited, Stuttgart, West Germany.

## We engineered the best-then improved it

We just weren't satisfied with the best when we engineered the Tannoy Variluctance Pickup Cartridge.

Our design engineers went to work right away and perfected the "Complidex," a brand new stylus assembly that utilizes with even greater efficiency both the cantilever and variable reluctance principles. Using a combination of two distinct metals our design engineers overcame the inevitable compromise between magnetic and mechanical requirements entailed by a conventional homogeneous material. Result-the new
"Complidex" Stylus has increased magnetic efficiency within the gap plus improved mechanical efficiency of the cantilever. Further development gives correctly graded damping without disturbing the optimum vertical-lateral ratio of compliance.
Like their predecessors, the new "Complidex" Styliwith either sapphires or diamonds-allow instantaneous replacement without tools. The new "Complidex" Styli can be used to convert the original (Mark I) cartridge to Mark II specification.

## SPECIFICATION

Each cartridge hand-made and laboratory tested Frequency response within 2 dB to $16,000 \mathrm{Kcs}$. No resonant peaks
No undamped resonances in sub-supersonic range Simple turn-over mechanism
Stylus assemblies completely independant
Instantaneous replacement of styli without
use of tools
Optimum lateral to vertical compliance ratio Very low effective dynamic mass
Output: 20 mV at 12 cm per second Termination load: 50,000 ohms. Tracking weight: 6 grams for all discs Available with either diamond or sapphire styli

## TANNOY

## TANNOY Mark II 'VARILUCTANCE' PICKUP CARTRIDGE

Tannoy Products Ltd. (Practitioners in Sound), West Norwood, London SE27. Telephone: Gipsy Hill II3I


The elements of ORYX instruments are mounted at the tip of the stainless steel shafts. No loss of heat and maximum efficiency are the result. Strictly controlled heat (not too little, not too much), giving exact soldering temperature. Robust construction without mica, ceramics, pins, etc. Simple, push-on spare blts, easily replaced. Finger-tip control.

## Model 6

Designed for soldering hair-springs, transistor work and miniature instruments. The only model with a non-replaceable bit. For 6 volts only.

## Model 6A

Designed for production and maintenance of hearing-aids, printed circuit-work and transistor-assemblies. For 6 volts only.

## Model 9

Designed for miniature radio and instrument work, relays, switches, small assemblies, etc. Available for 6 v ., 12 v . or 24 v .

## Model II

Designed for special high-temperature work, soldering temperature approx. $100^{\circ}$ above normal. For 6 volts only.

## Model 12

Designed for maintenance work on television, and radio sets, alrcraft work, electronic instruments. Available for 6 v ., 12 v ., 24-28 v. or 50 v .

## Model 18

Designed for high-speed soldering on production lines. Prevents fatigue by operators and damage to surrounding components. For 6 volts only.


Mod. Consumption Bit Dia. Weight Length Price Sp. Bits

| 18 | 18 watts | $\frac{3}{16} \mathrm{in}$. | $\frac{3}{4} \mathrm{oz}$. | $7 \frac{1}{4} \mathrm{in}$. | 35/- | 3/6 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12 | 12 watts | $\frac{3}{16} \mathrm{in}$. | $\frac{1}{2}$ oz. | 6 in . | 25/- | 21- |
| 9 | 8.3 watts | $5 \cdot 32 \mathrm{~mm}$. | $\frac{1}{4}$ oz. | 6 in . | 25/- | 21- |
| 6A | 6 watts | 332 in . | $\frac{1}{4} \mathrm{oz}$. | 6 in . | 25/- | 2/- |
| 6 | 6 watts | $\frac{1}{16} \mathrm{in}$. | $\frac{1}{4} \mathrm{oz}$. | 6 in . | 251- | - |
| 11 | 10 wates | 5/32 in. | $\frac{1}{2}$ Oz. | 6 in . | 35/- | 7/6 |

## (114)

## Teferadió 5A

## SIMPLE EFFICIENT

Made by AWA, Australia's largest manufacturer of telecommunications equipment of all types. AWA are approved contractors to the Crown Agents.
The 5A is in use by Governments and priv. ate networks in many places. Please write for details.


## SPEECH COMMUNICATION

The AWA Teleradio 5A is a low-power H.F. transmitterreceiver for distances up to several hundred miles over land or sea.
One to four channels between 2 and $9 \mathrm{mc} / \mathrm{s}$ may be pre-tuned in the crystal-controlled transmitter. The receiver also tunes from 550 to $1540 \mathrm{kc} / \mathrm{s}$.
The standard model operates from 12 -volt battery. An A.C.

* Regd. Trade Mark 34699 (Aust.) model is also available.
AMALGAMATED WIRELESS (AUSTRALASIA) LIMITED 47 YORK STREET, SYDNEY, N.S.W.

1,000 metre for professional use $1,200 \mathrm{ft}$. on plastic spool 600 ft . on plastic spool 300 ft . on plastic spool 150 ft . miniature tape for postal recordings.
Special tapes for GRUNDIG TK. 9 , TK. 819 and Stenorette.

Also
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The effect of different voltages on initial heating-up time is shown. Whilst 4 V is the standard voltage normally employed, 6 V will cause no harm, and accumulators are a useful source of current supply.

* Activated by light thumb pressure on the switch ring. When pressure is released, current is automatically switched off - thus greatly reducing electricity consumption, wear on copper bit and carbon element.
* Length, $10^{\prime \prime}$; weight, $3 \frac{1}{2}$ ozs. ; can be used on 2.5 to 6.3 volt supply ( 4 volt transformer normally supplied) or from a car battery.
* More powerful than conventional I50-watt irons; equally suitable for light wiring work or heavy soldering on chassis.
* Simple to operate ; ideal for precision work.
* Requires minimum maintenance - at negligible cost ; shows lowest operating costs over a period.

For full particulars, including guarantee terms and free trial facilities, please write to the sole concessionaires in this country :-
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## heats up from cold in 6 seconds!

Manufactured for Enthoven Solders Ltd., by Scope Laboratories, Melbourne, Australia.

Designed on an entirely new principle, this light-weight, versatile iron is eminently suitable for soldering operations in the radio, television, electronic and telecommunication industries. For test bench and maintenance work it is by far the most efficient and economical soldering iron ever designed. Ideally suitable for use with Enthoven Aluminium Cored Solder (melting point $260^{\circ} \mathrm{C} .500^{\circ} \mathrm{F}$.).

## Switch to the



Superspeark

## Soldering Iron

as being used by the
Royal Society Antartic Expedition for the International Geophysical Year.
 SONIXGRAM REGD.

A COMPLETE DOMESTIC HIGH FIDELITY SYSTEM

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    PABO-1 Tapo Pre-Ampana
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Collaro 2010 Transeription Gramophone Unit


Armstrong FM 56 Tuner - Permeability Tuning - Freedom from Drift - Magic Eye Tuning - 3 position HT Supply Sockel Automatic Limiting 222/1/-

## Armstrong AM44 Tuner

- Fariable Selectivity 4 marebands - Magic Eye Tuning Infinite Impedance Detector - Cathode follorer
Ouspui £19/17/-

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- 10-12 vatts uliraliuear 0.1\% total harmonic at 8 watto Within 1 dB $15 \cdot 30,000$ cps 4 inpus postlious Equaliser Filler Separate Bass and Treble controle


## WEYRAD I.F. TRANSFORMERS AND RATIO DETECTORS FOR AM AND FM.



## TYPE P.23/IR

COMBINED AM/FM TRANSFORMERS $470 \mathrm{KC} / \mathrm{S}$ AND $10.7 \mathrm{MC} / \mathrm{S}$. AM BANDWIDTH $8 \mathrm{KC} / \mathrm{S}$ AT -3DB. FM BANDWIDTH $200 \mathrm{KC} / \mathrm{S}$ AT -I. 5 DB .

## TYPE P.23/2R

COMBINED AM I.F. TRANSFORMER AND FM RATIO DETECTOR. THE R/D PROVIDES A PEAK-TO-PEAK BANDWIDTH OF $340 \mathrm{KC} / \mathrm{S}$ WHICH IS LINEAR OVER $220 \mathrm{KC/S}$.

BOTH ASSEMBLIES ARE MOUNTED IN SCREENING CANS SIZE I $\times 1 \frac{5}{8} \times 2 \frac{3}{8} \mathrm{in}$. THE TRANSFORMERS ARE DESIGNED FOR USE WITH EF89 AND EABC80 VALVES OR THEIR EQUIVALENTS. THE P.23/I AND P. $23 / 2$ CAN BE EMPLOYED WITH FRONT-END TUNERS INCORPORAT NG THE FIRST I.F. STAGE, ALTERNATIVELY WEYRAD P. $22 / 4$ FM I.F. TRANSFORMER CAN BE USED.

THESE COMPONENTS ARE USED IN OUR AM/FM RECEIVER-CONSTRUCTIONAL CATALOGUE OF INDIVIDUAL COMPONENTS 6d.

BOOKLET 2/6.
WEYMOUTH RADIO MANUFACTURING CO., LTD., CRESCENT STREET, WEYMOUTH, DORSET.


## $9 \times 5$ Elliptical Speaker Type 59T.

This unit is designed to give good quality at domestic volumes, it can be operated successfully in conjunction with a normally good receiver.

Where a Power Output Stage providing more than 4 watts is used, 2 or more speakers are recommended.

Flux Density, 8000 gauss ( 27,500 Maxwells) Frequency
Response 40-12,000 Cps.
RETAIL PRICE. . . . . . . . . . . 38/2 inc. Tax
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## "YOU CAN RRLY ON US"

Stockists of all Radio and Electronic components for manufacturers, laboratories, Educational authorities, and the amateur.

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[^12]
## GILSON TRANSFORMERS



SIDE-BY-SIDE COMPARISON IN OUR DEMONSTATION ROOM OF ACOS ACOUSTICAL - ARMSTRONG CHAPMAN - collaro connoIsseur GARRARD GOODMANS LEAK LENCO ROGERS R.C.A. TANNOY W.B. - WhARFEDALE.

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Full range of Transistors including all Components for Mullard 200 mW . Amplifier and TCC Midget Capacitors.
"Q-MAX" CHASSIS CUTTERS STILL the easiest and quickest way of curting holes in SHEET METAL.
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(b) $-1 / 6$; (c)-
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ARMSTRÓNG PB 409 AM/FMRADIOGRAM CHASSIS - 9 valves

- L., M., S., and F.M.

28 gns .

# memuRDO red range connectors FOR BACK RACK MOUNTING 

NOW AVAILABLE

16 WAY

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## GOLD PLATED CONTACTS

POSIITVE POLARISATION
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THAN -005 OHM

- CURRENT RATING - 5 AMPS

PER CONTACT

- MOULDINGS - RED NYLON LOADED P.F

INSERTION AND
WITHDRAWAL FORCE


Model SFB/3 (Req. Design 881557)

IMPEDANCE CURVE. Note the unusually level impedance which typifies the wide frequency response.

frequency in c/s

* ATTRACTIVE APPEARANCE * FREE-STANDING \& EASILY MOVED * RESONANGE FREE : SANDFILLED BAFFLE $\star$ OMNI-DIRECTIONAL * FREQUENCY RANGE : $30 \mathrm{c} / \mathrm{s}$ to $20,000 \mathrm{c} / \mathrm{s}$ * MODERATE PRICE

Here, at last, is a wide range, high quality speaker which can easily be accommodated in the average room, can be placed in the best position for listening, and be moved against a wall when not in use, with choice of Walnut, Oak, Mahogany and Maple Vencers.

> SPECIFICATION. Size $34 \mathrm{in} . \times 31 \mathrm{in}$. $\times 12 \mathrm{in}$. Weight 64 lbs . Impedance $8 / 15$ ohms. Bass Resonance $30 / 35 \mathrm{c} / \mathrm{s}$. Max. input 15 watts.
> UNITS. W12/SFB, 10in. Bronze/SFB, Super 3 . The 12 in . and 10in. units are in parallel. This arrangement gives very smooth results over the full range with a 3 DB gain at low frequencies. The Super 3 is again parallel via 4 Mfd. capacitor and is mounted on a small baffle facing upwards. The efficiency of the system is high and will give full domestic volume from any good 5 -wart amplifier. The baffle is sandfilled; there is no cabinet resonance because there is no cabinet.
> The 12in, and 10in, units are specially built and MATCHED FOR OPTIMUM RESULTS from this system. Baffles cannot be supplied separately.

Tropical model made with resin bonded plywood can be supplied at £2 extra Made and guaranteed by WHARFEDALE WIRELESS WORKS LTD • IDLE • BRADFORD • YORKS Telephone : Idle I235/6

## Don't Buy an Amplifier Until You have Heard


with the collaboration of one of Britain's foremost transformer designers Verdik have produced the finest instrument in its class measuring only $8 \frac{34^{\prime \prime}}{} \times 4 \frac{3^{\prime \prime}}{} \times 5^{\prime \prime}$ CONSIDERABLY SMALLER THAN THE SIZE OF THIS PAGE.

## SPECIFICATION

PRE-AMPLIFIER


## Direct readings of



New (4) 523B

## ELECTRONIC COUNTER

## Extreme dependability

## Etched, unitized circuits

Permits viewing time-interval start and stop points on oscilloscope

## - High accuracy crystal oscillator circuit

## A Trouble-localizer lights

- Counts pulses of selected voltage level

Construction of the new -hp- 523B is highest quality throughour. Etched circuits are rugged, ultra-dependable. Circuits are arranged for complete visibility. Trouble-localizer lights and plugs disconnecting circuit elements further simplify maintenance.

Exclusive features include a pulse output for oscilloscope Z-axis modulation permitting visual identification of the time-interval start and stop points on the input waveform measured. There is also a pulse count discriminator counting only pulses of voltage above a pre-determined level; and a high accuracy, high stability crystal controlled oscillator. Controls are color-coded, concentric, functionally arranged. Readings are direct in clear, bright numerals; decimal is automatic and illuminated.

The broad range and versatile usefulness of $-h p-523 \mathrm{~B}$ is indicated by the Specifications at right. Model 523B is designed for utmost speed and simplicity in measuring production quantities, rpm, nuclear pulses, power line frequences, repetition rates, time intervals, pulse lengths, shutter speeds, velocities, relay times, frequency ratios, phase delay, etc.

FREQUENCY

## TIME

 PERIOD
## 10 cps to 1.1 MC !

With transducers, $-h p$ - 523B also provides local or remote measurement of weight, pressure, temperature, acceleration, etc.

## BRIEFSPECIFICATIONS

FREQUENCY MEASUREMENT:

Range:
Accuracy:
Accuracy:
Input Minimum:
Input Impedance:
Gate Time:
Gate Time:
Reads Directly In:

10 cps to 1.1 MC
$\pm 1$ count $\pm$ crystal stability 0.2 v RMS

Approx. 1 megohm, $30 \mu \mu \mathrm{f}$ shunt
$0.001,0.01,0.1,1,10$ seconds
KC. Automatic decimal
KC.

## Range:

Accuracy:
Input Minimum:
Input Minimum:
Input Impedance:
Input Impedance:
Gate Time:
Standard Counting:
Reads Directly In:
0.00001 cps to 10 KC
$\pm 0.3 \%$ ( 1 period); $\pm 0.03 \%$ ( 10 periods)
Iv RMS
Approx. 1 megohm, $40 \mu \mu f$ shunt
1 or 10 cycles of uniknown
$10 \mathrm{cps}, 1 \mathrm{KC}, 100 \mathrm{KC}, 1 \mathrm{MC}$, Externa?
Sec, msec, $\mu$ sec; automatic decimal
TJME INTERVAL MEASUREMENT:

Range:
Accuracy:
Input Minimum:
Input Impedance:
Trigger Slope:
Trigger Amplitude:
Standard Counting:
Reads Directly In:
STABILITY:
DISPLAY TIME:
OUTPUTS:
PRICE:
$3.0 \mu \mathrm{sec}$ to $100,000 \mathrm{sec}$ ( 27.8 hrs )
$\pm 1 /$ std. freq. counted $\pm$ stability
1 I peak. dc coupled
Approx. 1 megohm, $25 \mu \mu$ shunt
Pos. or neg. on start/stop independent or common channels
-300 to +300 v adjustable
$10 \mathrm{cps}, 1 \mathrm{KC}, 100 \mathrm{KC}, 1 \mathrm{MC}$, External
$\mathrm{Sec}, \mathrm{msec}, \mu \mathrm{sec}$, automatic decimal $2 / 1,000,000$ per week. Also WWV Variable 0.1 to 5 sec , or indefinite Secondary standard: $10 \mathrm{cps}, 1 \mathrm{KC}$ rectangular, $100 \mathrm{KC}, 1 \mathrm{MC}$ sine wave. $\$ 1,175.00$

Data subject to change without notice. Price f.o.b. factory.
HEWLETT-PACKARD COMPANY
Represented by
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Telephone: Byron 8783



## The BTH Dual Channel Loudspeaker

To those striving for perfect sound reproduction, the new BTH K10A SPEAKER is an essential component. Designed to meet an exceptionally exacting specification, the unit is now available with or without cabinet to the keen enthusiast.
The equipment comprises an 18 -inch cone for low frequencies; with a co-axial pressure-type unit and horn for high frequencies, and is complete with cross-over network.

The matching impedance is 15 ohms ; the cross-overfrequency is 1700 c.p.s.; and the total weight of unit and filter is 38 lbs .
List Price (less cabiner) : $\mathbf{5 4 5}$


RATING . UP TO 2 kW

FREQUENCY RANGE . . . $2 \mathrm{Kc} / \mathrm{s}$ to $2 \mathrm{Mc} / \mathrm{s}$
H.F. power transformers of outstanding efflciency are the latest additions to the Mullard range of high quality components designed around Ferroxcube magnetic cores.
Utilising the unique characteristics of Ferroxcube to the full, Mullard H.F. transformers are smaller, lighter, and less costly than transformers using alternative core materials. These advantages are particularly marked in transformers required to handle powers of up to 2 kW , between the frequency range $2 \mathrm{kc} / \mathrm{s}$ to $2 \mathrm{Mc} / \mathrm{s}$.
Mullard transformers are already finding wide use in applications as diverse as ultrasonic H.F. power generators and aircraft power packs operating from an aircraft's normal A.C. supply. In the latter application, the low leakage field of Ferroxcube can eliminate the need for external screening, thereby reducing the size and weight of the transformer even further. As with all Mullard high quality components, these $H$.F. power transformers are designed and built to engineers' individual specifications. Write now for details of the complete range of components available under this service.

## Comuoisseurr

## 3 SPEED MOTOR

The turntable with a $4 \%$ variation on all three speeds.

The Connoisseur motor is made for the perfectionist. It is one of the finest turntables in the world.
The speed change is arranged mechanically and gives a 4 per cent variation on all speeds. A synchronous motor, which is virtually vibrationless with low noise level and hum induction, maintains a constant speed at all settings. There is no braking action to obtain speed change.

The 12 in . turntable is lathe turned in non-ferrous metal. The main spindle, which is precision ground and lapped to mirror finish, runs In phosphor bronze bearings.

A sound, precision engineering job, the Connoisseur motor provides the foundation for perfect reproduction.

Price E20, plus P. Tax $\mathbf{6 8 / L I / -}$


Matching Connoisseur Pick-up Mark II with a frequency range from 20-20,000 cycles:
Pick-up complete with I head fitted with Diamond armature 68/19/- plus P Tax 63/16/6.

## A. R. SUGDEN \& CO. (ENGINEERS) LTD.

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Phone: Halifax 69169. Grams: Connoiseur, Brighouse.

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# HGE TTEMPERATURE 



- For operation in ambients of up to $85^{\circ} \mathrm{C}$
- Elimination of cooling fins for compactness
- Light alloy construction for lightweight


## - High stability

are some of the salient features of this new range of rectifiers, primarily developed for Service requirements, but now available in production quantities for all applications. Units can be supplied with various finishes to comply with RCS. 1000 categories H.I and H. 2 .

Details of this new range of

## WESTALITE (4)

## RECTIFIERS


can be obtained from
Dept. WW.II. WESTINGHOUSE BRAKE \& SIGNAL CO., LTD., 82, York Way, King's Cross, London, N.t.

# QUESTION 

## ANSWER

## "Why don't dealers stock and recommend our Amplifiers and Tuners, etc?"

## "Because they cannot afford to, as we give their discount to YOU (the public)."

This direct trading explains why our products, though in the top class, are so much cheaper than our competitors.

## What we are and what we do.

Firstly we are quite large manufacturers af Audio Amplifiers, Radio Feeder Units, Portable Record Players, Speaker and Amplifier Cabinets and custom built Complete High Fidelity Radio and Record Reproducers.
Secondly we are Retailers of Gramophone Units, Autochangers, Speakers, Tape Recorders, etc., etc.
We recommend only that which we know to be of good performance and of sound construction. We are not in the group of traders who sell job lines at apparently low prices because they are obsolete or faulty. On the other hand our finances are such that we do not have to sell you an expensive article if we know that a less expensive unit will do your job perfectly. If any reader should have his mind set on a high-priced amplifier
of another make and would like to save money if possible, we should like to make the following clear-cut offer: If he buys one of our "Symphony" Model Amplifiers (Standard or Studio version) and is not entirely satisfied with it he may return it for full credit against any other amplifier or tuner on the market. It should be emphasised at this stage that we can supply any Amplifier, Radio Tuner, etc., advertised.
Our chief Engineer, who is operating a Technical Guidance Service, is available daily including Saturdays from 10 a.m. to 6 p.m. or will deal with enquiries by return of post.
Our new illustrated Catalogue and supplement will be a great boon to those desiring quality equipment for modest expenditure. Send two $2 \frac{1}{2}$ d stamps for your copy now. It may well save you pounds! All our equipment is on demonstration at our showroom in conjunction with a variety of Pickups, Speakers, etc. If you can possibly call we shall be pleased to see and help you.

The New No. I SYMPHONY" AMPLIFIER MARK III is a 3 -channel 5 -watt Gram/Radio Amplifier with astonishingly flexible tone control. You can lift the treble, the bass, or-and here is the unique featurethe middle frequencies to suit your own ear characteristics and the record or radio programme being heard. It is thus possible to arrange the frequency-response of the amplifier to a curve equal and opposite to the resultant curve of the other items in the chain so that what finally curve or the other items in the chain so that what finally
registers in the brain is as per original. This flexibility of control is even more important than the nominal linear response of the amplifier, as the pickup speaker, etc., are not linear. Independent Scratch-Cus is also fitted and special negative feed back circuit employed. The Amplifier can accommodate a wide variety of records from old 78 s to new L.P.s, and there is full provision for Radio Tuner, Tape take-off and Playback. It is available to match $2 / 3$ or 15 ohms speakers. Price 12 gns. (carriage 7/6). Fitted in portable Steel Cabinet, 2 gns. extra.

The New No. 2 "SYMPHONY" AMPLIFIER MARK III, as No. I but with 10 -watt Push-Pull triode output and triodes throughout. Woden mains and output transformers and choke. Output tapped 3, 7.5 and 15 ohms. Provision for Tuner and. Tape. Comperes with the most expensive amplifiers on the market yet cost ony 16 gns. (carriage 7/6). Fitted in portable Stee Cabinet 2 gns. extra.
"SYMPHONY" AMPLIFIERS WITH REMOTE CONTROL


## Remote Control Pane

Both the above model Amplifiers are available with all controls on a separate Control Panel with up to 4 ft . flexible cable which simply plugs into the amplifier. Enables the Amplifier proper to be sat in the bortom of a cabinet whilst the controls are mounted conveniently higher up. Extra cost 2 gns.

No. I "SYMPHONY:' F.M. TUNER. High grade Instrument with extremely silent background. Based on the latest type of permeabilityAssembly advanced design housed in anti adiation hroud giving hroud giving vity and high music/noiseratio Suitable for amplifiers in the highest fidelity class. $\mathrm{Cl} / 8 / 8 /=$ Power Pack 63/7/6. Magic eye 61 extra required.

N.R.S. EMPRESS FM/VHF TUNER/ADAPTOR fine little job, will plug into any radio and add F.M \&13/15/-. Magic eye assembly $\boldsymbol{\ell} 1$ extra if required. Ditto mounted in beautiful dark walnut cabinet complete with magic eye $16 \frac{1}{2}$ gns. Carriage $7 / 6$.
" SYMPHONY " AM/FM RADIOGRAM CHASSIS. Very high grade radiogram chassis combining Long.
M.H.F.bands. Large en. graved dial. Push/pull output for
high quality. high quality.
Complete complete
with $10 i n$. Goodmans 5 peaker
26 gns. Carriage and packing 7/6.

RECOMMENDED GRAMOP HONE UNITS All current. Collaro Units in stock for immediate delivery.

GARRARD TA \& TB 3/SPEED UNIT,
New model RC88 AUTOCHANGER
6900
RC98 with variable speed 21710
Variety of Pickup cartridges available in Garrard shell to fit TA, RC88 and RC98. Leaflets on Collaro and Garrard Gram. Units on request.

LENCO GL50, 4-speed continuously variable from above 78 r.p.m. to below 16 r.pm. Special Autostop. Price with Studio "O" or "P." head or Goldring variable reluctance head, E21/17/10.

LENCO GL55, as above but without pickup and autostop but fitted with Special Device for Groove Location and knob which completely disengages drive-wheel Suitable for use with any pickup, especially transcription cypes and B.I. Arm. Price $\mathbb{E} 17 / 10 / 4$. Immediace delivery guaranteed.

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LENCO GL56, as GL5S but with Studio or Goldring 500 pickup, £23/7/.
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ELECTRONICS, RADIO, TELEVISION

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NOVEMBER 1956

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[^17]A high gain transistor preamplifier is described which uses a single OC70 in grounded emitter connection. The supply voltage is 250 V and the load more than $330 \mathrm{k} \Omega$. The basic circuit arrangement is shown at ' A '.

The preamplifier has distinct advantages over a thermionic valve preamplifier. First, two valves would be required to get comparable gain. Second, thermal noise is sufficiently low for it to approach that of a valve circuit, and the transistor preamplifier is of course completely free from hum and microphony. The current drain is low, about 0.7 mA from the 250 V supply.

This preamplifier has comparatively high output impedance and best results are obtained when it is fed into the high input impedance of a thermionic valve amplifier. It can be used with low impedance microphones, pickups and tape heads. For a nominal circuit, with an input voltage of 5.5 mV and a gain of about 330 , the output voltage is about 1.8 V . The input impedance at $\mathrm{X}-\mathrm{X}$ is $200 \Omega$ and the output impedance at $\mathrm{Y}-\mathrm{Y}$ is of the order of $5 \mathrm{k} \Omega$. The frequency response depends to a certain extent on the source impedance. With a $50 \Omega$ source a 3 dB reduction in gain is reached at $15 \mathrm{c} / \mathrm{s}$ and $12 \mathrm{kc} / \mathrm{s}$.

Although designed for a 250 V supply voltage the preamplifier operates usefully with supply voltages down to as little as 100 V . It can be used successfully with a $30 \Omega / 50 \Omega$ microphone and a 100 V supply coming from a valve amplifier into which it feeds.

The preamplifier can be fed from the same supply as a valve amplifier by rearranging the circuit as at ' B '. In the setup shown at ' $B$ ' one side of the output is earthed, whereas in ' $A$ ' the input and output are 'floating' at some voltage above the chassis. The input is fed between base and emitter through R1 and C1 so that R5 does not contribute a.c. negative feedback but forms part of the load. The output in ' $A$ ' can be taken from between C2 and chassis if desired, but it then becomes slightly smaller than when taken from C2-C3, because R5 no longer forms part of the load. Although the input terminals in both ' A ' and ' B ' are 'floating', there should be no risk of hum being introduced provided the preamplifier is mounted reasonably close to the microphone or pickup.

If the circuit is used in arrangement ' $A$ ' care must be taken not to short the input terminals to earth. In the arrangement shown at ' $B$ ' the current through the pickup or microphone will not be excessive if the input is accidentally shorted to earth.

High voltage gain is obtained from a transistor in the same way as for a pentode valve, by operating it with a high load and feeding from a high supply voltage. In both ' $A$ ' and ' $B$ ' sufficient d.c. stabilisation is provided by the potential divider R2-R3 and emitter resistor R5 to ensure satisfactory operation up to an ambient temperature of $45^{\circ} \mathrm{C}\left(110^{\circ} \mathrm{F}\right)$. Besides the d.c. feedback provided by R2, R3 and R5, there is an a.c. feedback path formed by R2, R1 and the source impedance, all
in series. Part of the a.c. voltage developed across this potential divider is tapped off and fed back into the base. By including a $100 \Omega$ resistor R1 in the a.c. feedback path, effective feedback is ensured even when the source impedance is very small. Apart from improving the frequency response and reducing distortion, the a.c. negative feedback decreases the input and output impedances. If thought desirable a.c. feedback can be prevented by making R2 up of two nearly equal resistances and bypassing the common point to ground by a suitable capacitance, of the order of $0.5 \mu \mathrm{~F}$.

Even with a supply voltage of 250 V the circuit is so designed that the collector to emitter voltage never exceeds the d.c. voltage rating of $\mathrm{V}_{\mathrm{c}} \max =-5 \mathrm{~V}$ for the OC70 under the worst possible combination of conditions. The effect of resistance tolerances, supply tolerances, change of ambient temperature, and spread in transistor characteristics have all been considered in the design. In order to keep available the maximum voltage across the transistor, while allowing a tolerance of $\pm 10 \%$ on the supply voltage, the resistor tolerances must be within $\pm 5 \%$ to prevent the 5 V rating from being exceeded. High stability resistors should be used. For a nominal circuit the collector current is 0.7 mA and the collector to emitter voltage approximately 4V. An OC70 with a low current amplification factor $a^{\prime}$ and low leakage current $\mathbf{I}^{\prime}{ }^{\prime}(0)$ in grounded emitter causes the highest collector to emitter


B


The circuit is similar to that described by James J. Davidson in an article called 'High Gain Transistor Amplifier,' published in Audio, Ocrober 1955. voltage of 5 V . A transistor at the other extreme gives a collector to emitter voltage of about 0.8 V at the maximum ambient temperature of $45^{\circ} \mathrm{C}$, and allows an output of about 400 mV (r.m.s.).

If resistors of $\pm 10 \%$ tolerance are used, the collector to emitter voltage should be metered to ensure that it does not exceed 5 V . The performance is the same as for resistor tolerances of $\pm 5 \%$ except that possibly less collector voltage swing will be available if all the resistors have their lowest extreme value.


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## "BELLING LEE" NOTES

T.V.-V.H.F.-F.M.

To many layman this is all very confusing. Television and the new F.M. sound transmissions from Wrotham and elsewhere are all V.H.F. and all V.H.F. transmissions are subject to multipath or reflected signals. In the case of television the result is multiple or ghost images due to the reflected signals arriving a split second after the direct signal, and the most likely cure lies in the use of a correctly sited directional aerial.
Now in the case of F.M. reception reflected signals are just as likely to manifest themselves but in this case they result in a form of distortion in speech or music which is sometimes just as difficult to eradicate as are the ghosts in television.
It is most unfortunate that so much was written in the early. days of F.M.' stressing the fact that only primitive aerials were required. It is a fact that F.M. can be picked up over astonishing distances either with a "built in" aerial, or with an added piece of wire. But remember, reflected signals are just as common as on television and where they occur there is distortion, so there is not the slightest sense in assuring yourself that because it is F.M. it must be perfect.
Whereas a vertical dipole is omnidirectional a horizontal one is directional at right angle to its plane. This helps, but get it high up, preferably out of the house and on a chimney for best results. It may be necessary to use a two or even a three-element array before the programme really comes in as the B.B.C. would like you to receive it.
Please ask us for a copy of the B.B.C. V.H.F./F.M pamphlet; it is well worth reading.


Advertisement of BELLING \& LEE LTD.
Great Cambridge Rd., Enfield, Middx. Written 20th Sept., 1956

## announce



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 Collapsable band Iarray

A recent addition to the "Belling-Lee" range of band $I$ aerials, the " $X$ " will find favour with the many members of the trade and the public who have asked for an aerial of this shape as an alternative to the Junior " H " mentioned below.
The new " X " is built to the usual high "Belling-Lee" standards. It has been developed from the original "Belling-Lee" "V" type array produced in 1946, and in fact consists of two "V's" placed "back-to-back".
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The "belling:LEE" Junior "H" retains its position as one of the most popular aerials for use in "close-in" situations in normal service areas, giving a slightly wetter performance than the " X ".

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WHAT IT COSTS. The average cost of an installation with 50 receivers would be under $£ \mathrm{r}, 500$ including the cost of the loop. The receiver incorporates four transistors and is powered by a single cell. Since the quiescent current is less than 0.5 m.a. it will only cost a few shillings a year to run each receiver-considerably less than any other electronic system.

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* The total hum and noise at $7 \frac{1}{2}$ inches per second $50-12,000$ c.p.s. unweighted is better than 50 dbs .
$\star$ The meter fitted for reading signal level will also read bias voltage to enable a leve! response to be obtained under all circumstances. A control is provided for bias adjustment to compensate low mains or ageing valves.
* A lower bias lifts the treble response and increases distortion. A high blas actenuates the treble and reduces distortion. The normal setting is inscribed for each instrument.
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* A heavy mu-metal shielded microphone transformer is built in for $15-30$ ohms balanced and screened line, and requires only 7 micro-volts approximately to fully load. This is equivalent to 20 ft . from a ribbon microphone and the cable may be extended 440 yds. without appreciable loss. * The .5 megohm input is fully loaded by 18 millivolts and is suitable for crystal P.U.s, microphone or radio inputs.

The amplifier, speaker and case, with detachable lid, measures $8 \frac{1}{4} \mathrm{in} . \times 22 \frac{1}{2} \mathrm{in} . \times 15 \frac{3}{4} \mathrm{in}$. and weighs 30 lb .

PRICE, complete with WEARITE TAPE DECK ... ... ... ... ... 884 0 0

* A power plug is provided for a radlo feeder unit, etc. Variable bass and treble controls are fitted for control of the play back signal.
* The power output is 3.5 watts heavily damped by negative feedback and an oval internal speaker is built in for monitoring purposes.
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One milliwatt output on 600 ohm line (.775V) for an input of 30 micro-volts on $7.5-30 \mathrm{ohm}$ balanced input.
Output balanced or unbalanced by internal switch. The meter reading is obtained by a valve voltmeter with I second time constant, which reads programme level, and responds to transient packs.
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# The <br>  <br> BAND III Convertor 



The Wolsey Convertor is a self-contained unit complete with power pack, comprising transformer with tapped primary for mains voltage selection, together with metal rectifier and R.C. smoothing.

Though employing the popular valve line-up of cascode coupled P.C.C. 84 and P.C.F. 80 as mixer and output amplifier, the circuit otherwise departs completely from the conventional.
Aerial input circuit has an input of 80 ohms over the Band III range this being achieved by critical resistance damping of the input transformer, from which the signal is fed to a neutralised cascode coupled R.F. amplifier. The cascode potentlal of this valve being adjustable so as to afford gain control.
Coupling between the cascode and the mixer grid of the oscillator section of the P.C.F. 80 is made by a choke capacity arrangement which is constant over the whole of the Band III range.
The oscillator section employs turret tuning of the eight channels and careful selection of component valves ensures minimum of drift.
I.F. output from the pentodè section of the P.C.F. 80 , which is of course the local B.B.C. frequency, is fed via a stepped-down transformer to the low impedance output socket.
Gain from input to output in excess of 20 dB may be easily obtained and no Convertor is passed on test unless it has at least this gain.

# OLYMPIC HONOUR FOR BRITAIN 


#### Abstract

We are proud to announce that our TL,/12 amplifiers have been chosen for use at the 1956 Olympic Games to be held in Australia.


It was in 1945 that H. J. Leak revolutionised the performance standards for audio amplifiers by designing the original "Point One" series, and we became the first firm in the world to market amplifiers having a total distortion content of 0.1 per cent. This claim was received with incredulity, but it was subsequently confirmed by the National Physical Laboratory and since then hundreds of TL/12 amplifiers have been used by the B.B.C., and Commonwealth and foreign broadcasting authorities, and thousands have been used by recording studios, leading musicians and musiclovers throughout the world. We were the only British exhibitor at the world's first Audio Fair which was held in New York in 1949 and the volume of our exports to the United States of Arnerica has grown steadily since then.

Further development woork resulted in our producing, at a much lower price but with the same high performance standards, the TL/10 amplifier. The TL/10 amplifier and "Point One" pre-amplifier received such an excellent reception when they were first exhibited at the Audio Fair in New York in October, 1953, that we received an initial order for 1,000 sets. Since then several thousand sets have been sold throughout the world. The output of the TL/ 10 is ample for highfidelity home music systems, and the quality of reproduction obtained is equal in every respect to that of the $T L / 12$. We always use the $T L / 10$ amplifier and " Point One" pre-amplifier for our public demonstrations of high-fidelity reproduction of gramophone records and radio. The TL/10 amplifier, when used with the best available complementary equipment, gives to the music-lover a quality of reproduction unsurpassed by any equipment at any price. Even when the complementary equipment falls below that of the best obtainable, the use of these amplifiers will enable one to obtain very marked improvements in reproduction.


Reprints of " The Gramophone:', article (May, 1955) by H. J. Leak, summarising his work and findings on Electrostatic and Dynamic Louidspeakers, are available on request, free of charge.

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## Make LEAK equipment the heart of your Hi-Fi system

Illustrated above is the LEAK TL/IO AMPLIFIER $£ 17.17 .0$ and " POINTONE " PRE-AMPLIFIER, $£ 10.10 .0$. These prices are made possible only by world-wide sales.

## SPECIFICATION:

## Circuitry

A triple loop feedback eircuit based on the famous TL/I2. The output transformer is the same size as in the $T \mathrm{~L} / 12$.
Maximum power output: 10 watts.
Frequency Response: $\pm 1 \mathrm{db} 20 \mathrm{c} / \mathrm{s}$ to $20,000 \mathrm{c} / \mathrm{s}$.
Harmonic Distortion: $0.1 \%, 1,000 \mathrm{c} / \mathrm{s}, 7.5$ watts output.
Feedback Magnitude: 26 db , main loop.
Damping Factor: 25.
Hum: -80 db referred to 10 watts.
Loudspeaker Impedances: 16 ohms, 8 ohms, and 4 ohms.

## " POINT ONE " PRE-AMPLIFIER

The handsome gold escutcheon plate contributes to the elegant appearance and blends with all woods.

## $\star$ Pickup

The pre-amplifier will operate from any pickup generally available in the world. A continuously variable input attenuator at the rear of the pre-amplifier permits the instantaneous use of crystal, movingiron and moving-coil pickups.

## * Radio

The radio input sockets at the rear permit the connection of the LEAK V.S. zuner unit. An input attenuator is fitted. H.T. and filament supplies are available from the pre-amplifier.
$\star$ Distortion
Of the order of $0.1 \%$.

* Hum

Negligible, due to the use of recently developed valves and special rechniques. $\star$ Input selector
Radio, tape, records: any and all records can be accurately equalised.

## $\star$ Treble

Continuously varíable, +9 db to -15 db at $10,000 \mathrm{c} / \mathrm{s}$.
t Bass
Continuously variable +12 db to -13 db at $40 \mathrm{c} / \mathrm{s}$, at $40 \mathrm{c} / \mathrm{s}$,
t Volume Control and Switch
The switch controls the power 'supply to the TL/10 power amplifiers. * Tape Recording Jacks

An exclusive feature. Readily accessible jacks are provided on the front panel for instantaneous use with Tape Recorders which have built-in (low level) amplifiers.

## OTHER LEAK PRODUCTS

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BAND III PRE-AMP


BAND III AERIALS


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bracket ..................
3 element array with cranked
mast and chimney lashing equipment $\qquad$
5 element array with swanneck mast and "U " bolt clamp for fitting existing mast from $\frac{1}{4}$ in to 2 in . dia... 5 element array with cranked mast and chimney lashing tquipment
8 element array with swanneck mast and "U" bolt clamp for fitting to tin. to 2in. dia. mast

## T.R.F. CONVERTER

New this month is a converter for T.R.F. Set Viewmaster-Electronic Engineering, etc. Small mods. to the TV are necessary as this must be turned into a superhet to stop re-radiating. Price complete with two valves $56 / 10 /$ -assembled-ready Oct./W:W.

THE "CRISPIAN"

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$16-50$ metres Short wave $187-550$ metres medium wave. 900-2,000 metres long wave. Flywheel Tuning.
Negative Feed Back.
Valves 6BE6, 6AT6, 6BW6, $6 \times 4$.
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Comprising the weil-known BC625 and BC624A. Units complete with 17 valves topes: $2-832,3-13 A 6,3-12 S G 7,3-9003,9003,4 \mathrm{G6G}$, 12J5GT, 12AF7GT, 12C8, 6S87. The Complete unit is in very good condition having very useful parts including Relays, Trans Cormers, Condensers, etc.
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Containing VCE97 with Mu-Metal Screen. 21 Valves: 13-EF50, 4-\$P61, 3-EA50. 8-EB34. Plus Pote., Switche 3, H.V. Cond., Reslstors, Muilrhead S/M Dini. Double Deek Chassis and Crrstal. BRAND NEW
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Size $15 \mathrm{~m} \times 8 \mathrm{in} . \times 2 \mathrm{n}$. Complete with $45 \mathrm{Me} / \mathrm{s}$. Pyo Strip, 12 Faves, 10 EF50, of Kesistors and Condeneers. Ner condition. vodification data supplied.

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 Unit contains VOR517 Cathode Ray 6in tube, complete with $\mathrm{Mu}-$ Metal screen, 3 EF50, $4 \$ P B 1$ and 1 SU4G valves, 9 wire-woind volume controls and quantity of wound volume controls and quantity of resistors and condenters. Offered BRAND NEW (less relay) at $67 / 6$ Plus $7 / 8$ carr.
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Complete with 11 valves 8 -SP61 SU4G, VO120. $\nabla$ R.92. As specified for thexpensive In $\begin{aligned} & \text { In ab }\end{aligned}$
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176
$\begin{array}{ll}15 & 0 \\ 10 & 0\end{array}$

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Valves $6 \mathrm{~K} 8,6 \mathrm{~K} 7,6 Q 7,6 \mathrm{~V} 6$ and $6 \times 5$. For A.C. Mains 200/250 v. Chassls size $13 \frac{1 n}{} \times 5 \operatorname{in} . \times 2$ in. Dial eize: $10 \mathrm{in} . \times 4 \frac{1}{2} \mathrm{in}$ Assembly is simplified by use of ready assembled coll pack. Illustrated bookle with full assembly instructions, with itemised price list, is $1 / 6$ post iree-or the ki


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Based on the booklet by Data Publications Litd., 21 - port free, including our individually priced Parts List. Highly senaitive, free
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 Biit into a strong wooden cabbet 15in. $\times$14in. $\times$ 9in. Completo with headphone 14in. $\times$ 9in. Complete with headphones
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No 38 TRANSMITTER/RECEIVER WALKIE-TALKEE. Range approx. ${ }^{5}$ milles Coverage $7.4,9, \mathrm{c} / \mathrm{z}$. 5 he set only good condition. We also have a vailable " 38 " sets absolutely complete with Jusction box, aerial, headibhonfs, microphoge, syake set of valven, with full
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 $\begin{array}{llllll}\text { ECC83 } & \text { O/ } & \text { ELA1 } & \text { 10/6 UP41 } & 10 / 6 \\ \text { EOC84 } & 15 /- & \text { ELS4 } & 11 / 6 & \text { UL41 } & 10 / 6\end{array}$


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We carry comprehensive stocks of all Band III Convertors by leading manulacturers, low-loss co-axial cable at 10 d . per yard. Let us have your enquiries. Any branded convertor supplied on E.P. terms!
MAINS TRANSFORMER BARGAINS. Limited quantities. Manufacturer plue $350-0-350,80 \mathrm{~mA}, 6.3$ v. 3 an, 5 v. 2 a . Half shrouded, dropthrough, $14 / 6$ 30080 mA .6 .3 ₹. 3 a 208. npat, $300-0^{-}$ ised drop-through tyjue, $8 / 8$ only ised drop-through type, $9 / 6$ only, plus $1 / 6$
$P$. \& $P$. Input $110 / 230 \mathrm{v}$. Auto load 230 v . 750 . Input 1107230 v . Auto load $230 \vee$. ment winding 6 v .3 a., $15 \vee .3$ a., 21.5 v . .6 a., aiko 5 v .2 a. Tropicallsed drop-
 200/250
plus $1 / 6 \mathrm{P}$. Input unversal mounting $16 / 6$,

GARRARD RC. 110 ilt 1 s-8peed mixer auto-changer unit with G.C. 2 t/o crystal a sealed manufacturers cartons with tting and operating instructions. 27/19/6 plus 5/- p. and p.
RECORD PLAYER CABINETS-to suit illypes of single record and auto-changer fully illustrated $45 /$. Send stamp


Anothes Cabinet Bargain 1 Speciß purchase intended for use in projection T.V. Easily recognised as betng of leading Hight Quality manufacturer's stock. Can the easily adapted to house tape recorder, amplifer, radiogram, etc. etc. Measurements oxternal 24 in. $\times 16 \mathrm{ha} \times 29 \mathrm{in}$. The whole is
mounted on castors. Unrepeatable baryalin at $£ 5 / 18 / 6$, plus $10 /$ - packing and carriage. at $85 / 18 / 6$, plus $10 /$ - packing and carriage.
We have $a$ large selection of all type cabinets A stamp will bring list


## EQUIPMENT



All for A.C. MAINS 200-250v., $50 \mathrm{c} / \mathrm{s}$,
Assembled 6 v . or 12 v. 4 amps. Fitted Ammeter and variable charge sel ector. Also selecto plug for 6 v . or 12 v. charging. Double fused. Well ven tilated steel case with blue hamme finish.
Ready for $69 / 6$ use with mains and output leads. Carr 3/9.

## SELENIUM RECTIFIERS

L.T. Types

$2 / 6$ v. $\frac{1}{2}$ a.h.w. 1/9 | $2 / 6$ |  |  |
| :--- | :--- | :--- |
| $6 / 12$ v. $\frac{1}{2}$ | a.h.w. | $1 / 9$ | F.W. Bridge Types F.W. Bridge Types 120 v. 40 mA . $6 / 12$

$6 / 12$
v..$~$

2 a. . | $6 / 12$ | v. 3 a. | $\ldots$ | $12 / 9$ | 250 v. 150 mA. | $9 / 9$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $6 / 12$ | v. 4 a. | $\ldots$ | $14 / 9$ | 300 v. $275 \mathrm{~mA} .12 / 11$ |  | CO-AXIAL CABLE. 75 ohms, tin., 8 d. yard. Twin screened feeder, 11d. yard.

5 CORE FLEX. Henleys circular rubber 14/36. Each lead colour coded. 1/6 yd.
DIAL BULBS, M.E.S., 8 v. 0.2 a., 6/9 doz. 6.5 ष. 0.3 a., $6 / 9$ doz.

ELECTROLYTICS (current production).
Tubular Types
$88 \mu \mathrm{~F} 450$ v Types
$88 \quad 1 / 9 \quad 8 \mathrm{mfd} .600 \mathrm{v} . . \mathrm{v}$
$8 \mathrm{mfd}, 500 \mathrm{v} . \quad 2 / 6 \quad 16 \mathrm{mfd} .350$
$16 \mu \mathrm{~F} 350 \mathrm{v}, \ldots \quad 2 / 3 \quad 16 \mu \mathrm{~F} 450$ ₹.
$16 \mu \mathrm{~F} 450 \mathrm{v} . .2 / 9 \quad 32 \mu \mathrm{~F} 350$
$16 \mu$ F 500 v. . $\quad 3 / 9 \quad 32$ mfd. 450 v.
$32 \mu \mathrm{~F} 350 \mathrm{v} . . . \quad 3 / 9$
32 mfd . $500 \mathrm{v} . \quad 5 / 9$
$8-16 \mu$ F 500 v. $4 / 11$
$25 \mu \mathrm{~F} 25$ v. .. $1 / 3$
$50 \mu \mathrm{~F} 12 \mathrm{v}$. $\quad . \quad 1 / 3$
$50 \mathrm{mfd}, 25 \mathrm{v} . . . \quad 1 / 9$
$100 \mathrm{mfd} .12 \%$ 1/9
$100 \mathrm{mfd} .25 \mathrm{y} . \quad 2 / 3$
$6,000 \mathrm{mfd} .6 \mathrm{v} .3 / 11$
Many others in stock.
VOLUME CONTROLS with long spindles, all values, less switch, $2 / 9$; with S.P. switch, 3/9. EX GOVT. METAL BLOCK PAPER CONDENSERS
2 mfd. $500 \mathrm{v} . . \quad 1 / 9 \quad 8$ mfd. $500 \mathrm{v} . . .4 / 11$ $\begin{array}{llll}4 \mathrm{mfd} . & 1,000 \mathrm{v} . & 3 / 9 & 8-8 \mathrm{mfd} .500 \mathrm{v} . \\ 5 / 11\end{array}$ $\begin{array}{lllll}4 \mathrm{mfd} . \\ 4 \mathrm{mfd} . & 1,500 \mathrm{v} . & 5 / 9 & 8-8 \mathrm{mfd} .500 \mathrm{v} . & 5 / 21\end{array}$ 4 mfd .400 v . plus $2 \mathrm{mfd} .250 \mathrm{v} . \ldots . .$. . $1 / 11$ EX GOVT. VALVES. VR137, EA50, EB34, 11d.; SP61 2/3; 4SHA 1/3; EL32 3/9; VS110 1/11; MU14 7/9; KT44 4/9; 615 3/9; 6V6G,
EX GOVT. UNITS, type RDF1 in original sealed cartons with 14 valves including $5 Z 4 \mathrm{G}$, etc., trans., L.F. choke, Rectifier, etc.s etc. We cannot enter into correspondence regarding these units which represent a really exceptional bargain at 29/9. Carr. 7/6.

OII FILLED BLOCK
CONDENSERS
Bryce $11-7$ mfd. 500 v . New unused Govt. surplus, only $5 / 9$ each.

T.R.F. RECEIVER A design of a 3 valve $200-250$
v. A.C. Mains v. A.C. Mains selenium rectifier. For inclusion in cabinet illustrated above or walnut veneered type. It
employs valves $6 \mathrm{~K} 7, S P 61,6 \mathrm{~F} 6 \mathrm{G}$, and is specially designed for simplicity in wiring, Sensitivity and quality is well up to standard. Point-to-Point wiring diagrams, instructions and parts list, 1/9. This receiver can be built for a maximum of $£ 4 / 19 / 6$ including cabinet. Available in brown or cream bakclite, or Avalable in bro
vencered walnut.

EX GOVT. MAINS TRANSFORMERS All $230 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$ input $120-0-120$ v. 40 mA . $120-0-120$ v. 40 mA.
$300-0-300$ v. $150 \mathrm{~mA} . \mathrm{m}^{2}$.
3 $300-0-300$ v. 150 mA., 4 v. 3 a. ....... 5/9 $250-0-250$ v. 80 mA ., 6.3 v. 3 a., 6.3 v. 1 a. Potted $4 \pm-3 \frac{1}{2}-3 i n$. 460 v. $200 \mathrm{~mA} ., 6.3$ v. 5 a. ............ 22/9 0-16-18-20 v. 35 a., 79/6. Carriage 7/6 extra. MANUFACTURERS SURPLUS

## TRANSFORMERS

Fully shrouded upright. Primary 200-230250 v . Sec. $425-0-425 \mathrm{v} .150 \mathrm{~mA} .6 .3 \mathrm{v}$. 3 a., 5 v, 3 ang 33/9. Clamped type 325-0325 v. $100 \mathrm{~mA}, 6.3$ v. $3.5 \mathrm{a}, 5 \mathrm{v}$ v. 2 a, Wearite $19 / 9$. $230-0-230 \times$ v. 60 mA .,
6.3 v. 2.5 a. Midget, $2 \frac{1}{2} \times 3 \times 2 \frac{1}{2}$ in. approx., 6.3 v. 2.5 a. Mi
$11 / 9 . \quad$ Post $2 / 9$.

## EX GOVT. SMOOTHING CHOKES

250 mA ., $5 \mathrm{H} . \mathrm{H}^{2} 50$ ohms
250 mA ., $3 \mathrm{H} ., 50 \mathrm{ohms}$
$150 \mathrm{~mA} ., 10 \mathrm{H} ., 50 \mathrm{ohms}$
$150 \mathrm{~mA} ., 6-10 \mathrm{H} ., 150 \mathrm{ohms}$, Tropicalised $100 \mathrm{~mA} ., 10 \mathrm{H} ., 100$ ohms, Parmeko 100 mA ., $5 \mathrm{H}, 100$ ohms, Tropical ised. $50 \mathrm{~mA} ., 50 \mathrm{H} ., 1,000$ ohms L.T. type 1 amp., 2 ohms

## SPECIAL OFFERS. Small 2 gang variables .0005 mfd, , $4 / 9$. $8-8 \mathrm{mfd}$., 450 v Electrolytics (midget) in lots of six, 1/6 ea

## R.S.C. BATTERY TO MAINS CONVERSION UNITS

Type BM1. An all dry bat-
tery eliminator. Size $5 \frac{1}{z}$ $4 \frac{1}{2} \times 2 \mathrm{in}$. approx. Completely replaces batteries supplying 1.4 v . and 90 v . where A.C. mains $200-250$ v. $50 \mathrm{c} / \mathrm{s}$. is battery portable receivers battery portable recelvers This includes latest low This includes latest low consumptiontypes, Complete kit with diagrams $39 / 9$, or ready for use, $46 / 9$

SILVER MICA CONDENSERS. 5, 10, 15, 20, 25, 30, 35, $50,100,120,150,180,200,250,300,330,400,470,500,1,000$ pfd. (.001 $\mu \mathrm{F}$ ), . 002 mfd . ( $2,000 \mathrm{pfd}$.). .005 mfd ., .01 mfd . All at 6 d . each, $3 / 9$ doz. one type.
VIBRATORS. Oak 2 v. 7 pin, synchronous, 7/9.


Type BM2. Size $8 \times 5 \frac{1}{2} \times$ 2 tin. Supplies 120 v., 90 v. and $60 \mathrm{v}, 40 \mathrm{~mA}$. and 2 v 0.4 a. to 1 amp. fully smoothed THEREBY COMPLETELYRE PLACING BOTH H,T. BATTERIES AND L.T. 2v. ACCUMU LATORS when connected to A.C. mains supply $200-250 \mathrm{~V}$ $50 \mathrm{c} / \mathrm{s}$. SUITABLE FOR ALL BATTERY RECEIV ERS normally using 2 v accumulator. Complete kit with diagrams and instruction $49 / 9$ or ready for use $59 / 6$.
T.V. CABINETS. For 15, 16 or 17 in . tube, Table model with doors, 79/6 carr. $7 / 6$.

## R.S.C. TRANSFORMERS

## FULLY GUARANTEED. INTERLEAVED AND IMPREGNATED

## MAINS TRANSFORMERS

Primaries 200-230-250 v. $50 \mathrm{c} / \mathrm{s}$
FULLY SHROUDIFD UPRIGHT MOUNTING
$250-0-250$ v. $60 \mathrm{~mA} ., 6.3$ v. 2 a $_{\mathrm{t}} 5$ v. 2 a.
Midget type, $2 \frac{1}{2}-3-3 i n$.
$350-0-350$ v. $70 \mathrm{~mA}, .6 .3$ v. 2 a., 5 v. 2 a.
$250-0-250$ v. $100 \mathrm{~mA}, 6.3$ v. 4 v. 4 a., c. 0-4-5 v. 3 a
250-0-250 v. $100 \mathrm{~mA} ., 6.3$ v. 4 a, 5 v. 3 a 250-0-250 v. $100 \mathrm{~mA} ., 6.3$ v. 6 a., 5 v. 3 a. for R1355 conversion
300-0-300 v. $100 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a}, 5 \mathrm{v} .3 \mathrm{a}$ 300-0-300 v. 100 mA., 6.3 v. 4 v. 4 a., c.t. $300-0-300$ v. 3 a.
$0-4-5$ v.
350-0-350 v. $100 \mathrm{~mA}, 6.3$ v. 4 a., 5 v. 3 a. $300-0-300 \mathrm{v} .130 \mathrm{~mA}, 6.3 \mathrm{v}, 4$ a., c.t., 6.3 v 1 a., suitable for Mullard 510 Amplifier $350-0-350 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .=4 \mathrm{v} ., 4 \mathrm{a} .$, c.t., 0-4-5 v. 3 a.
$350-0-350$ v. 150 mA., 6.3 v. 4 a, 5 v. 3 a. 350-0-350 v. $150 \mathrm{~mA} ., 6.3 \mathrm{v} .2$ a., 6.3 v 2 a., 5 v. 3 a.
425-0-425 v. $200 \mathrm{~mA}, 6.3$ v. 4 a., c.c. 6.3 v. 4 a. c.t., 5 v. 3 a., suitable 450-0-450 v. $250 \mathrm{mLA}, 6.3$ v. $6 \mathrm{a}, 6.3$ v

6 a., 5 v. 3 a,
TOP SHROUDED DROP-THROTGF TIP 250-0-250 v. $70 \mathrm{~mA}, 6.3$ v. 2.5 a $260-0-260$ v. $70 \mathrm{~mA} ., 6.3$ v. 2 a., 5 v. 2 a $350-0-350$ v. $80 \mathrm{~mA} ., 6.3$ v. $2 \mathrm{a}, 5 \mathrm{v} .2 \mathrm{a}$ $250-0-250$ v. $100 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 5 \mathrm{v} .3 \mathrm{a}$ 300-0-300 v. 100 mA., 6.3 v. 4 v. 4 a., c.t., 0-4-5 v. 3 a.
$350-0-350$ v. $100 \mathrm{~mA} ., 6.3 \mathrm{v} .4 \mathrm{a}, \mathrm{c}, \mathrm{L}, 5 \mathrm{v}, 3 \mathrm{a}$. 350-0-350 v. 100 mA., 6.3 v. -4 v. 4 a c.t., 0-4-5 v. 3 a.
$350-0-350$ v. $150 \mathrm{~mA}, 6.3$ v. 4 a., 5 v. 3 a E.H.T. TRANSFORMERS, $2,500 \mathrm{v}$ $5 \mathrm{~mA}, 2-0-2$ v. 1.1 a., $2-0-2$ v. 1.1 a. $5 \mathrm{~mA}, 2-0-2$ v. $1.1 \mathrm{a} .$,
for VCR97, VCR517
$25 / 9$
$23 / 9$

P9
13/9
$16 / 9$
18/9
22/9
$23 / 9$
$22 / 9$

## FILAMENT TRANSFORMERS

Primaries $200-250$ v. $50 \mathrm{c} / \mathrm{s}$
$\begin{array}{llll}6.3 \text { v. } 1.5 & \text { a. .. } & 5 / 9 & 0-2-4-5-6.3\end{array}$

| .3 v. 1.5 a. | 5/9 | 0-2-4-5-6.3 v. |
| :---: | :---: | :---: |
| 6.3 v. 2 a. | 7/6 | 4 a, |
| 0-4-6.3 v. 2 a. | 7/9 | 6.3 v. 6 a. |
| 6.3 v. 3 a. | 8/11 | 12 v. 3 a. or |
| 12 v .1 a . | 7/9 | 24 v. 1.5 a. |

$\frac{12 \mathrm{v} .1 \text { a. ... } 7 / 9 \quad 24 \mathrm{v} .1 .5}{\text { CHARGER TRANSFORMERS }}$
Al with 200-230-250 v. $50 \mathrm{c} / \mathrm{s}$. Primaries $0-9-15$ v. $1 \frac{1}{6}$ a., 11/9; 0-9-15 v. 3 a., 16/9 $18: 9: 0-9-15$ ४ 5 a, 19/9: 0-9-15 マ. 6 a. 23 a.

## ELIMINATOR TRANSFORMERS

Primaries 200-250 v. $50 \mathrm{c} / \mathrm{s}$.
120 v. 40 mA. 5-0-5 v. 1 a.
90 v. $15 \mathrm{~mA}, 96-0-6$ v., 250 mA
14/9

## OUTPUT TRANSFORMERS

26/9 Midget Battery Pentode 66:1 for 3S4, etc.
33/9 Small Pentode 5,000 $\Omega$ to $3 \Omega$
Standard Pentode, $5,000 \Omega$ to $3 \Omega$
Multi-ratio 40 mA. $30: 1,45: 1,60: 1$,
90:1, Class B Push-Pull
49/9 Push-Pull 8 w 6 , $9 / 6$
$69 / 6$ Push-Pull 10-12 Watts 6 V 6 to $3 \Omega$ or $15 \Omega \quad 15 / 9$
Push-Pull $10-12$ Waits to match 6 V 6 to
3-5-8 or $15 \Omega$
Push-Pull 15-18 W atts, sectionally wound
Push-Pull 10-15 Watts, Ultra Linear,
designed for Mullard Amplifier 510 .
Push-Pull 20 Watt high-quality sectionWilliamson fype exact to spec.
SMOOTHMNG CHOKES
$23 / 9250 \mathrm{~mA}, 5 \mathrm{H}, 100$ ohms
$150 \mathrm{~mA}, 7-10 \mathrm{H} ., 250$ ohms
100 mA ., $10 \mathrm{H}, 200$ ohms
$80 \mathrm{~mA} ., 10 \mathrm{H} ., 350$ ohms
$36 / 660 \mathrm{~mA} ., 10 \mathrm{H},$.400 ohms

## R.S.C. A6 ULTRA LINEAR 30 WATT AMPLIFIER

## NEW 1956 DESIGN. HIGH FIDELITY

 PUSH-PULL UNIT EMPLOYING SIX incorporated. Sensitivity is extremely high incorporated. Sensitivity is extremely high. for full output. THIS ENSURES THE SUKABILITY OF ANY THP OR MAKE OF MICROPHONE OR PICKUP. Separate Bass and Treble controls give correction for long playing records. ANOUTPUT SOCKET WITH PLUG IS OUTPUT SOCKET WITH PLUG IS INCLUDED FOR SUPPLY OF 300 m . and $6.3 \geqslant 1.5$ a. FOR A RADIO FEEDER UNIT. Price in kit form with easy-to-follow wiring diagrams. Only GNS. Or Factory built with 12 months ${ }^{\circ}$ guarantee 50/- extra. TERMS ON ASSEMBLED UNITS with extra input. DEPOSIT 28/9 and 9 monthly payments of 28/9. If control can be provided so that two separate inputs such as "mike" and glam., etc. etc., can be simultaneously applied for mixing purposes. Extra cost of this $13 /-$. Cover as illustıated $17 / 6$ extra.


Type 807 output valves are used with High Quality Sectionally wound output transformer specially designed for Ultra Linear operation. Negative feedback of 17 D.B. in main loop. CERTIFIED PERFORMANCE EXPENSIVE UNITS AVAII. MOST EXPENSIVE UNITS AVAIL. ABLE. Frequency response $\pm 3$ D.B., $30-20,000$ c/cs., 12 D.B. "lift" at $50 \mathrm{c} / \mathrm{cs}$.,
12 D.B. "lift" at $12,000 \mathrm{c} / \mathrm{cs}$. , Hum and 12 D.B. "lift" at 12,000 c/cs., Hum and noise $70 \mathrm{D}, \mathrm{B}$. down. Good quality reliable
components used. Chassis finish blue components used. Chassis finish blue crackle. Overall size $12 \times 9 \times 9 \mathrm{in}$, approx. Power consumption 150 watts. For A.C. mains $200-230-250 \mathrm{v} .50 \mathrm{cjcs}$. Outputs for 3 and 15 ohm speakers. EQUALLY SUITABLE FOR THE CONNOISSEUR OR FOR LARGE HALLS, CLUBS, or OUTSIDE FUNCTIONS. IDEAL FOR USE WITH MUSICAL INSTRUMENTS SUCH AS STRING BASS, ELECTRONIC ORGAN, GUITAR, etc. FOR DANCE BANDS, GARRISON THEATRES, etc. etc.
We can supply Microphones, Speakers, Rotary Converters, etc., at keen cash prices or on terms with amplifiers.
EXPORT ENQUIRIES INVITED



GARRARD 3-SPEED AUTOMATIC RECORD CHANGERS Latest Model. Type RC110. Fitted hlgh adelity turnover number. Brand new cartoned. Only $£ 7 / 19 / 6$, plus $3 / 6$ carriage.
IIM.V. LONG PLAYING RECORD TURNTABLE COMPLETE WITH CRYSTAL PICK-UP (SAPPHIRE STYLUS) f3/19/6 (approx. halt price). Carr. 5 f- (for $200-250$ Only A.C. Maina).

MICROPEONES. High fidelity crystal types. Acos 33-1 hand or desk type, $50 /$ - Plezzo with heary, floor base and telescople stem, 26/19/6.
R.S.C. 4-5 WATT HIGH GAIN AMPLIFIER tYPE A5

vill long playing record equallsation. Hum level is negll gible, beling $71 . \mathrm{D}, \mathrm{B}$. down, $15 \mathrm{D} . \mathrm{B}$. of negative feedback fo used. H.T. of $300 \mathrm{v}, 26 \mathrm{~mA}$. and L.T. of $6.3 \mathrm{v}, 1.5 \mathrm{a}$. is available for the supply of a Radio Feeder Unit or Tape Deck pre-amplifier. For A.C. miains Input of $200-230-250$ r., $50 \mathrm{c} / \mathrm{s}$. Output for $2-3$ ohm speaker. Chassis is not nlive Kit is complete in every detail and includes fully punched chassls (with baseplate) with the blue hammer Anlsh, and value at oniy $84 / 15 /$-, or assembled ready for usc $25 /-$ extra, plus $3 / 6$ carrlage.
R.S.C. A7 3-4 WATT QUALITY AMPLIFIER A hlghly sensitive 4 -valve amplifier using negative feedback and having an excellent frequency response. Pre-amplifier and Tone Control stages are incorporated with separate Bass and Treble controls siving full tone compensation for Long Playing records. Suitable for siny kind of pick-up neluding latest high fidelity types. H.T. of 250 V .20 mA , Unit, etc. ONLY 40 millivolte Input required for full output. Fully isolated chassls with baseplate. For A.C. mains $200-260$ v. 50 cycles, Output for $2-3$ ohm speaker. Complete kit of parts with point-to-point wiring diagrams and instructions. Only $£ 3 / 15 /$, carr. $3 / 6$ or factory buil $2^{2 / 6}$ extra.


BRAND NEW B.S.R MONARCH 3-SPEED MIXER erystal pick-up and dua point papphire atylij for chandard or long playing records. Plays tea 7in., 10in.
or $12 i n$. intermixed. or 12in. intermixed.
A.C. maius $200-250$ A.C. maine $200-250$
c/ca. Supplied in Bealed
cartons with operating ivstructions, £7/15/- plus $3 / 6$ cart

QUALITY AMPLIFIERS. $3-\frac{4}{4}$ watt, speclally designed for use with BSR or Garrard changers. Latest mindature type valves used. Separate Bass and Treble Contruls and
Vol. Control with mainsswiteh. For A.C. 200-250 v. mains Ready for use, only $£ 3 / 19 / 6$, carr. $3 / 6$.
Elliptical $7 \times 4$ in. with 2.3 ohme speech equpment
COLLARO HIGH FIDELITY MAGNETIC PIOK-UPS High Impedance, brand new, boxed at frwetion of norme price. Only 31/\%.
P.M. SREAKERS, $2-3 \mathrm{ohm}$, bin. Goodmans, 17/8, 64 in R.A., 15/9, 8in. Rola, 19/9, 10in. R.A. 26/9, 12in. Plesser 29/11.

## PLESSEY DUAL CONCENTRIC

 12 in. P.M. SPEAKERS
(15 ohms), consisting of a high quality 12 in . speaker, of orthodox design supporting a small elliptical speaker ready wired with choke and condensers to act as tweeter. This high fidelity unit is highly recommended for use with our A8 or any similar amplifier. Rating is 10 watts. Price only £5/17/6.

## Radia Sumply Co. (leeds) lto.

## 32 THE GALLS. - LEEDS, 2.

Terms C.W.O. or C.O.D. No C.O.D. under E1. Postage $1 / 9$ extra on all orders under 62, 2/9 extra under 55 unless carriage charge stated. Full Price List 6d. Trade List 5d. Open to Callers: $9 \mathrm{a} . \mathrm{m}$. to $5.30 \mathrm{p} . \mathrm{m}$. Saturday until $1 \mathrm{p} . \mathrm{m}$.

## R.S.C. ULTRA LINEAR I2-WATT AMPLIFIER



NEW 1956 MODEL A8 HIGH-FIDELITY. PUSH ${ }^{-}$ PULL AMPLIFIER WITH "BUILT-IN" IONE CONTROL, PRE-AMP. STAGES
High sensitivity. Includes 5 valves ( 807 outpats), High Quality sectionaly wound output transformer, speceially de-
signed for Ultra Linear operation, nnd rellable small conFOR A A current manufacture. INDIVIDUAL CONTROLS FOR BABS AND TREBLE " LIft" and "Cut." Frequency regronse $\pm 3 \mathrm{db} .30 \cdot 30,000 \mathrm{c} / \mathrm{cs}$. Six negative feerlbick loopas.
 makes and types of pick-ups and practicatly all microvhones Comparabie with the very beat designs. For STANDARD or LoNg PLAYING RECORDS. For MUSICAL INSTRUMENTS such as STRING BASS, GUTTARS, ete. OUTRUT SOcKET with plug provides 300 v . 20 ma, and 6.3 v. 1.3 a.
 for 3 and 15 ohm mains $200-230-230$. 60 c/cs. Output Chaseis is fully punched. Full instructions and point-topoint wiring diagrams supplied.
Unapproachable value at $57 / 15 /$ or factory built 451- extra. Carrlage 101-. If required lourred metal cover
with 2 carring handles can be supplied for $17 / 6$. Where an extrai input socket with associated volume control in required for mixing purposes this can be provided for 13/extra. TERMS ON ASSEMBLED UNTTS with extra input as mentioned above. DEPOSIT 25/6 and nine monthly payments of 22/4.
ROTARI CONVERTERS. 200 watte. Lnput 12 v. D.C. Output 230 ซ. 50 c/es. A.C. Only 7 gns. Carr. 7/6.
deflant record playing turntable complete Wire MAGNETIC PICK-UP. Pick-up is high impedance ype. Unit ss housed in a beautiful wainut veneered ot attractive design. For all standard records (78 r.p.m.).
Limited number. Brand new cartoned, $£ 5 / 17 / 6$, carr. $7 / 6$. W.B. "STENTORIAN " HIGE FIDELITX P.M. SPEAKERS. RF1012, 10 watti. 16 ohm (or 3 ohm) speech coll. Where a heally good quality speaker at a low priczis requoridance, 24/10/9. Plense state whether 3 ohm or 15 ohm required.

## SUPERHET FEEDER UNIT

Design of a high quality Radio Tuner Unit (specially suitable 2or use writh any of our Amplifers). A Triode Heptode Detector Delayed A.V.C. Ae./Orid.F/C Coupling is by botiom end Condenser Coupling giving freedom from alignment troubles when Ae. of varylng lengths and capracity are used. Both Frequenoy Changers and I.F. valves are A.V.C. controlled from the very low distortion Double Diodea so arriuged that very high Percentage modulation of the Transmitter can be handled without distortlon. The Feed for the delayed A.V.C. is arranged so that A.V.C. distortion is avoided. The W. Ch. Sw. Incorporates Gram.
position. Controls are Tuning, w., Ch., and Vol. Outnut wotll logd most Ampliffers requiring 500 M . V. input dependlug on Ae. looation. Onty 250 v. 15 mA . H.T., and L.T. of 6.3 ₹. 1 amp . required from amplifier. Size of unit approx. 9.8-6.7in. high. Bend 8.A.E. for illustrated leaflet. Total
bullding cobt is $£ 4 / 15 /-$. Point -to point wiring dig bullding coot is $£ 4 / 15 /$-, Point to point wiring diagram ind instructions, $2 / 6$.


## HIGH FIDELITY FOR THE HOME CONSTRUCTOR

## TUNING UNITS RADIORECEIVERS

COMPLETE KITS
The MULLARD ${ }^{5} 510$ '

sign and needs no recommendation from us. Our Kit is complete to Mulland's specification, lncluding the latest GILSON ULTRA LINEAR OUTPUT TRANSFORMER and the entire MULLARD Valve line up. ALL SPECIFLED
COMPONENTS are suppled. PRICE OF COMPLETE KIT OF PARTS $811 / 11 / 0$
| STERN'S "fidelity " PRE-AMPLIFIER TONE CONTROL UNIT
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Comprises the Marn AMPLIFIER of very popular
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have modified with the ex$\begin{array}{cc}\text { press } \\ \text { of its } & \text { purpose } \\ \text { use }\end{array}$ of its use
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A 2 -stage high amplifer having SEPAR. ATE BASS and TREBLE CONTROLS and deaigned to give up to approx. ${ }^{\text {w }}$ watth with very pleaing
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$\left\{\begin{array}{c}\text { SPECIAL OFFER!\}} \\ \text { GOODMANS I2in. } \\ \text { AUDIOM } 50 \\ \text { P.M. SPEAKERS } \\ 10 \text { watts. Limited number } \\ \text { only. Listed at } f 6 / 15 /-. \\ \text { LASKY'S PRICE } 97 / 6 \\ \text { Post free }\end{array}\right\}$ COLLARO RC54. With Studio O t.o. crystal plck-up. 28.18.6 Carr. 5/
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Brand new and unused. A.C./D.C. 200/230 volts. 1.F. $465 \mathrm{kc} / \mathrm{s}$. A.V.C., 4 watts output, 3 -station pre-set, frume aerial, fully
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$13 / 6$
$10 / 6$
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allowance of $4 / 6$ will be made if the mask is not required.
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to order. NOTE SIZES:
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Frame or line blocking osc. transFocus Magnets Ferrox-dure P.M. Focus Magnets, Iron Cored $300 \mathrm{~m} / \mathrm{s}$ Smoothing
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Frequency $95-150 \mathrm{Mc} / \mathrm{s}$ ( 2 to 3 metres) Valve line-up: 1st and 2nd R.F. Amp. VR. 136 (EF.54), 1st Local Oscillator VR. 65 (SP. 61 ), 2 Oscillator Multipliers VR. 136 (EF.54). 3 I.F. Amp. VR. 53 (EF.39). A.G.C. 6Q7. Output 6J5. Muting
 VR. 92 (EA.50). Noise Limiter VR. 92 (EA.50). B.F.O. 6J7. Mixer VR. 136 (EF.54). De Mod. 6 Q7.
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Drop thro' $350.0-350$ v. 70 mA .6 v . 2.5 amp .5 v. 2 amp., $14 / 6$.

Chassis mounting or drop thro'. Pri, $110-150 \mathrm{\nabla}$. Sec. $350-0-350,250 \mathrm{~mA}$.,
 C.T., 0.5 amp
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Chassis mounted and fully shrouded. $80 \mathrm{~mA} .6 \mathrm{v} .3 \mathrm{amp} ., 5 \mathrm{v}, 2 \mathrm{amp} ., 14 / 6$. Drop thro 270-0-270 $60 \mathrm{mAA}, 6 \mathrm{v}$. 3 amp ., 11/6.
250 v. $350 \mathrm{~mA} ., 6.3$ マ. 4 ang twice 2 2 а., 19/6.
Auto-trang. Output $200-250$ H.T. 500 v. 250 mA., 6 v. 4 a., twice, 2 จ. 2 a., $19 / 6$.
Auto-trans. Input 200-200. H.T. 350 v. 350 raA . Beparate L.T. 6.3 จ. 7 a., 6.3 v. if amp., 5 v. $3 \mathrm{amp} ., 25 /-$ P. \& P. 3/-.

Mains transformer, fully impregnated. Input $210,220,230,240$. Sec. $350-0-350$ 100 mA . with separate heater-transformer. Prl. 210, 220, 230, 240 вес. 6 amp, and $5 \mathrm{~F}, 2 \mathrm{amp} ., 30 /-\quad$ P. $\& \mathrm{P}$. 6 amp
$350-0-35076 \mathrm{~mA} .6 .3$ v. 3 a. $\operatorname{tap} 4 \mathrm{v}$. 6.3 จ. 1 я. $13 / 6$.

500-0-500 195 m 4 ₹ C.T 4 \& 4 C.T. 4 a., 4 v. C.T. 2.a a., $27 / 6$.

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64 in . M.E. Speaker. 1,000 ohm field, 15\%.
R. \& T.V. Energised 6lin. Spesker, with O.P. trans, feld coll 175 ohm 9/6. P. \& P. 2/6.
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5/-. With trimmers, 7/6.
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7,000
and
14,000,
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\section*{ADVANCE CONSTANT VOLTAGE} TRANSFORMERS. Input \(195-250\) volts, 50 cy cles, output 230 voles. 50 cycles, 150 watts Brand new E7/101. each.

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\section*{POWER UNIT TYPE 285} 230 volts 50 c.p.s., mains input. Outputs E.H.T \(2 \mathrm{kv} .5 \mathrm{~m} / \mathrm{a} . \mathrm{H} . T .350\) voles \(200 \mathrm{~m} / \mathrm{a}\). L.T. 6.3 voles 17 amps E.H.T. and H.T. supplies fully smoothed, double choke and paper condensers, etc., and complete with rectifier valves 5 U4G, VUli20 an EF50. A genuine bargain at only 59/6 each.
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Ideal for tape recorders, amplifiers, etc., extra sensitive. Supplied in'good condition and tested, only \(4 / 6\) each. Size only zin. square.

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Fit this new convertor not to your set but inside your set, even 9 in . table models, and retain that professional look.
This convertor has been evolved since the I.T.A. transmissions began, and is based upon experience gained in the conversion of very many Band I sets in the London area.
IT will convert any set, any age, TRF or Superhet
IT includes station switching
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IT is totally screened
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CONVERTOR wired and aligned with fitting instructions
KIT complete in every detail, less knobs
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CIRCUIT and instructions in detail (free with kit)
BAND III AERIALS (send torlist), from.
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Erie resistor kit, \(17 / 4\); Erle 1 meg. pots, \(4 / 6\) each; TCC condenser klt, \(55 /-\); PAR TRIDGE components with loose lead terminations (inoludes packing charge), Maine trans., 65/6; ohoke, 34/6; output trans.. 85/6; W.B. components. Choke \(18 / 8\) output trans., 32/-; Denco drilled chassia, 14/8; De
COMPONENTS are still available for the following F.M. TUNERS:-Wireless World,
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12 v. inpur., outpur 360 v. \(30 \mathrm{~m} . \mathrm{a}\). Cont. or 310 v. \(70 \mathrm{~m} . \mathrm{a}\). intermittent.
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Ceramic 5 PIN valveholders for bors it-

EiLECTRON MULTTIPLIERS Troe PIA Oily 35/ ea. \(0^{2}{ }^{2}\) tot
CRYSTAL HAND MICRÖPHONES with polishod grille and handle, four foot of screened lead. Only \(21 /=\). P. \& P. I/WORTH DOUBLE
POWER UNITS IN BLACK METAL CASE. 2001 260 v. input, \(200 / 250\) v. \(60 / 80\) ma. output, fully smoothed and filtered, also gives 31 v. D.C. and 6.3 v. 3 a. A. C., fitted with \(6 \times 5\) rectifier. Only 50/-ea. Carr. paid.
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RACK MOUNTING panels an Ig. long br
 ABSORPTION WAVEMETEES. 3 to \(3 . \mathrm{me}_{\mathrm{C}}\) in 3
 TRANSMITTTER TUNING CONDS. GY Jommon,
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For Cathode Ray Tubss having Heater/Cathode short tor C.R. Tubes with lalling emission Type A. Low leakago win


Ditto with msins primaries \(12 / 6\) each.
Type B. Mains input \(220 / 240\) volts. I ov Capacity. Multi Output 2, 4, 6.3, 7.3, 10 and 13 volts. Input has two tips Which incresse outpat volts by \(25 \%\) and \(50 \%\) respectively. This transformer is suitable
With Tas panel \(21 /\) esch.
Type C. Low eapacity wound translormar for use with 2 volt Tabes with falling emission. Input \(220 / 240\) volt \(17 / 8\) each. and clearly marked with selevant data.
NOTE:-It is essential to use mains primary types with
T. \(\%\). receivers having series connected heaters. RESISTORS. All values. 10 ohms to 10 meg., \(\frac{1}{6} \mathrm{~W} ., 4 \mathrm{~d}\);
 100 ohrms to 10 meg. 5 Watt \(\left.\begin{array}{l}\text { I0 watt } \\ 10 \text { wati } \\ \text { WIRE-WOUND RESISTOR } \\ \text { watt }\end{array}\right\} \quad 25\) ohms \(-10,000\) ohms
 WTRE-WOUNDPOTS. \({ }^{3}\) WATT LAB COLVEINE, ETC.
 Knurled
All
Slotited
25
Khms
 O/P TRANSFORMERS. Heav Values, 100 K., 8/6.
\(5 / 6 ;\). 100 .
W/W EXT. SPEAKER \(5 / 6 ; 100\) EXT., 8/8.
WPEAKER
CONTROL \(10 \Omega ~\) CONTROL \(10 \Omega / 3 / \stackrel{\text { Multi- }}{ }\) ratio. push pull, 6/6. Tapped small peutode, 3/9. Hygrade P CEOKEO 1510 . mA. 11/8: \(20 / 15 \mathrm{H} .120 / 150 \mathrm{mA}\). , 5/-; 25/20 H. 100/120


\section*{1.F. TRANSFORMERS \(7 / 6\) pair}
\(465 \mathrm{Kc} / \mathrm{s}\) Slug taning Ministure Can, \(2 h \times\) Iin. \(x\) in. High \(Q\) and good band width. By Pye Radio, Data sheet supplied.
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ALADDN FORMERS and cores. tin., \(8 \mathrm{~d} . \mathrm{im}\) in. 10 d . ch Can. and core. in. aq. \(\times 11 \mathrm{ln}\). and Sin. Eq. × 2hin., 2/- ea. Epicyclic ratio 8:1, 2/3.
SLOW MOTLON DRIVES. EM, TYANA. Midget Soldering Irou. \(200 / 220 \mathrm{v}\). of \(230 / 250 \mathrm{vo}\) SOLON MDDGET IRON. 25 w., 24/-.
MIEE TRANS. Ratio \(60: 1,3 / 9 \mathrm{es}\)., new and boxed. MKE TRANS. Ratio \(60: 1,3 / 9\) es., new and boxed.
MAINS DROPPERS. \(3 \times 1 \frac{1}{5} \mathrm{in}\). Three Adj. SIIders, .3 amp . 750 ohms, \(4 / 3.2\) amp., 1,000 ohms, \(4 / 3\). 100 ohms LINE CORD. \& amp., 80 ohms, per Ioot, 2 amp., 100 ohms her loot, 2 . way 6 d . per loot, 3 way 7 d . per foot.

\section*{CRYSTAL MIKE INSERT by Acos Precislon
Price 6/6 \\ \(1: \times 3 / 6 i n\).
Bargain.
o transformor
required. LOUDSPEAKERS P.M. 3 OHM \\ 5in. R.A., \(17 / 6\). \\ 7in. \(\times\) 4in. Goodmans, 21/. 81n. Elac., \(22 / 6\) 10is. R.A.. 30/-. in. M.E. \(2 . \overline{\mathrm{k}}\). or 2 k . fleld, tapped \(0 . \mathrm{P}\). trañé., \(24 / 6\). TSL TWEETER LSH75, 12/6 \\ CRYSTAL DIODE G.E.C., 2/- GEX34, 4/CRYSTAL SET CONSTRUCTION, \(1 /-1 / 2 / 6\). EANDBOOK OF GERMAMUM CIRCUITS, 2/6. SWITCR CLEANER Fluid, squirt sponit. \(4 / 3\) kin, TWIN GANG TUNING CONDENSERS. .0005 mid , midget less trimmers, \(6 / 6 ; .0005\) standard size with trimmers and
feet, \(8 /=\) : less \(\mathrm{trimmers}, 8 /-\) ditto, soiled, \(2 / 6\). \(3 \rightarrow\) gang Ieet, \(8 /-\) les
500 le}

SUPERHET COIL PACK \(27 / 6\).
 thou dlagram and circuit.
alve holdders. Par int. Oet., 4d. Ef50, EA50, \(8 d\) g12A, CRT, 1/3. Eng. and Amer, 4, 5, 6, 7 and 9 pin, \(1 /\) MOULDED Marda and Int. Oct. ©d, B7G, B8A, B8G, B9A. \(9 \mathrm{d}\). B7G with can, 1/6, VCR97, 2/6. B9A with can, \(2 / 6\). CERAMC. EF50, B7G, Int. Oct., \(1 / \%\) By with can, \(1 / 9\).
2i- Spin. Baffe silk covered, 1/- esch
WAVECHANGE SWITCHES
2 p. 2-way, 3 p. 2 -way, short spindle
5 p. 4 -way, 2 water, long spindle
2 p. B-way, 4 p. 2-way, 4 p. 3-way, long spindle
KNOBS. GOLD ENGRAVED. Walnut of Ivory, 11 in,; diam., \(1 / 6\) each. "Pocus," "Contrast," "Brilliance,"
"Briliance-On/Or"" "On-Of" "Yolume" "Vol "Brillignce-On/Orr", "On-OR," "Volume," "Vol- "Treble" "Bass,"
On-0ff," "Tone," "Tuning," "Tren "Warvechange", "Radio Gram," "SM.L. Gram,"
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THREE WAVEBANDS S.W. \(16 \mathrm{mu}-50 \mathrm{~m}\). LW, \(800 \mathrm{~m} .-2,000 \mathrm{~m}\).

FIVE VALVES \(4 W, 800 \mathrm{~m} .-2,000 \mathrm{~m} . \quad\) ECE 42, EF41, EBC41 12 month Guarantee. A.C. 200/250 v. 4-way switch. Short-Mediam-Long-Gram. A. V.C. and Negative
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MATCEMD SPEAKERS FOR ABOVE CRASSIS Sino, 19/6; 10in., 25/=; 12in., 30/-.
RECOMMENDED FOR ABOVE CHASSIS

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Collaro Autochanger RC53I for 78 R.P.M. 10 in , and 12 in . Records. Brand new in maker's boxes! High impedance, lightmatch any amplifier or radio. S5.19.6
Less than half price.
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TERMS: Deposit \(£ 4\) and 6 monthly payments of \(15 /\). Brand new Plessey 3 -speed Antochange Mizer Uni for 7, 10 and \(12 i n\). Records. Twin Hi-Fi Xbel Eleai with Duopoint sapphire otylus. Plays 4,000 records Sprung 5 in. Depth 2in. Super Quality. Post A.C. \(200 / 250\) -

Walnut Veneered Playing Desk cat out ready for B.S.R. MONARCE. 3-speed Motor and Turntable
with eelecting switch for 33,45 and 78 8.p.m. recorde. with selecting switch for 33,45 and 78 8.p.m. records. MONARCE Lightweight Pick-up with Acos Xital turnover head, separate Sapphire stylue for I, P. and standard recot
£ \(4 / 15 / 6\), post \(2 / 6\).
AMPLION. 3-speed Single Record anit with Acos 37 records. Starting Switch Aatomaticslly places Piok-up on records, 7 jn ., 10 in . of 12 in . Auto Stop. Baseplate \(12 \times 8 i \ln\). Height 21in. Depth 18in. Price £\%/15/6.

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Ready wound coils, two EF80 valves, all components, punched chassis, circuit diagram wiring plans. COMPLETE KIT for mains operation. \(200-250\) v. A.C. 63/10/-.
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TAPEMASTER RECORD HEAD, \(45 /-\)


Will amplify \(\begin{aligned} & \text { T.V. PRE-AMP. (McMICHAEL) }\end{aligned}\)
Will amplify outpot of your Band 3 Conrerter. Tun
ablo Channels 1 to 5 . Midget size. HIgh gain fring ablo Channels 1 to 5 . Midget iize. High gain fring
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 SPECLAL MAINS POWER PACK for above, 25/- oxtra

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Long spindles. Guaranteed 1 year. All values
10,000 ohms to 2 Meg. No 8 w. S.P.Sm. D.P.Sw Lin or Log Tracks

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Semi-air spaced Polythens insulated ing. dis. Stranded
core. Ideal Band III. Losses out \(50 \% \ldots .9 \mathrm{~d}\). yd. STANDARD
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\begin{tabular}{llcc}
\hline COAXIAL PLUGS & \(1 / 6\) & DOUBLE SOCKET & \(1 / 3\) \\
SOCKETS & \(1 /-\) & OUTLET BOXES & \(4 / 6\)
\end{tabular}
Band a Twies rif Band 3 Dipole indoor, 6/6. Band 3 Twies, 7/6 pr.
BALANCED TWIN FEEDER per yd. \(6 \mathrm{~d} .80 \Omega\) or 300 D TWIN SCREENED BALANCED FEEDER 1/- V O 3000 TRIMMERS, Ceramic, 30, 50, 70 pl., 9d, \(100 \mathrm{pt} ., 150 \mathrm{pt}\) \(1 / 3.250 \mathrm{pf}\)., \(1 / 6.600 \mathrm{pl}\)., \(750 \mathrm{pf} ., 1 / 9\).
ALUMINIUM CEASSIS. 18 s.w.K. Plain, untrilled, with 4 sides, riveted corners and lattice fixing holes, \(6 / 8 ; 13 \times 9 \mathrm{in}\)., \(8 / 6 ; 14 \times 11 \mathrm{in}\) 。, \(10 / 6 ; 15 \times 14 \mathrm{in}\)., \(12 / 6\) and \(18 \times 16 \times 3\) in., \(16 / 6\).
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MULTICORE SOLDER 60/40. 18 s.w.e., \(3 \mathrm{~d} ., 18 \mathrm{~s} . \mathrm{w} . \mathrm{g} ., 4 \mathrm{~d}\). Yd.

\section*{PURETONE RECORDING TAPE, \(12 / 6\)}
\(1,200 \mathrm{ft}\). on standard Aitting \(7{ }^{7-}\) Plastio reels. Spools \(5^{\circ}\) mefal \(1 / 6,7^{\circ}\) plastic \(4 / 3\).
SENTERCEL RECTIFIERS. E.H.T. TYPE FLY-BACK VOLTAGES. K3/25 \(2 \mathrm{kV}, 4 / 3 ; \mathrm{K} 3 / 403.2 \mathrm{kV}\)., \(6 /-\mathrm{K} 3 / 45\), MAINS TYPE, RM1, 125 v.; 60 mA., \(5 /-\); RM2, 100 mA.
 G.E.C. Neom Osglim, M.B.C, 180 volt, \(2 / 6\).
coris. Wearite, " \(P\) type, \(3 /-\) each. Oemor Midget "Q" type adj. dust core, \(4 /-\) each Ahl rangec- \(3 / 8\). TELETRON, L. \& Med., T.B.F." with reaction, \(3 / 6\).
E.F. CHOKES, mon cored. 14 M.F., \(3 /=\) each.
F.M. TUNER COIL KIT, 22/6. H.F. eoll, Aerial coll, \(\begin{array}{ll}\text { Obellator coil, two I.F. Transformers } & 10.7 \mathrm{Mc} / \mathrm{s} \text {. Detec- } \\ \text { cor transformer and heater choke. } & \text { With circuit and }\end{array}\) or transformer and heater choke.
component list, using four 6AM6.
CONDENSERS. New stock 001 mld 7 kV, TC.C.5/8 Ditto 20 kV ., \(9 / 6 ; 100 \mathrm{pl}\). to 500 pl . Micas, \(6 \mathrm{~d} . ;\) Tabular 500 v. . 001 to \(0 / 01\) mfd., \(8 \mathrm{~d} .05,1,1 /=; .25,1 / 6 ; .5,1 / 8\);
 SLLVER MICA CONDENSERS. \(10 \%\), 5 pf. to 500 pf . \(1 /-600 \mathrm{pf}\). to \(3.000 \mathrm{pf}, 1 / 3\). DITTO \(1 \% 1.5 \mathrm{pf}\). to 500 pf .
NEW ELECTROLYTICS. FAMOUS MAKES.


FULL WAVE BRIDGE SELENIUM RECTIFIERS. 2. 6 or

for charging at 2 . 6 or 12 F . 1 amp., 13/6; 4 amp.. \(21 /\). All BERNARDS books in stock.
VALVE MANUALS 1 \&
\(\mathrm{II}, 5 /\) ench part.
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Paoked in metai case \(7 \times 1\) itn. dia., \(4 / 6\).
TOGGLE SWITCKES.
S.P.D.T. (ex Govt.), each 1/-
NUTS, BOLTS AND WASEERS

\begin{tabular}{|c|c|c|c|c|}
\hline A SMALL & SELE & ON FI & M OUR & stocks \\
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\]} & \multicolumn{2}{|l|}{New \& Guaranteed} \\
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6ALs} & 1016
584 \\
\hline & 5/6 & & & \({ }^{524} 12 \mathrm{AT} 7\) \\
\hline 174 & 9D2 & \(2 / 6\) & 6 K 0 & B8C41 \\
\hline 185 & EF50 & \(2 \times 2\) & 6K7G & EBF80 \\
\hline 3S4 & Equip. & E1148 & EB91 & ECE35 \\
\hline 3 V 4 & SP61 & EB34 & HVR2 & EOL80 \\
\hline \(5{ }^{514}\) & \({ }_{8} 8\) P42 & 3/6 & (bear) & ECH43 \\
\hline 6AM8 & EF92 & 3D6 & EF36 & EF41 \\
\hline 6AT6 & & 6H6M & & EF80 \\
\hline 6 K 8 & 7/6 & \(7 / 6\) & \(7 / 6\) & ELAl \\
\hline 68.47 & 6BE6 & 6V6G & EL32 & KT33C \\
\hline 6SN7 & 6BW6 & \(6 \times 4\) & HVRPA & MU14 \\
\hline 6V6GT & 6F6 & \(6 \times 5\) & PEN25 & 1207 \\
\hline EBC33 & 6K6GT & 807 & U22 & PY81 \\
\hline EFF50 & 6K7M & EF39 & VP23 & 3524 \\
\hline Sylv. Red EF91 & & \[
\frac{11 / 6}{Y_{51}-U:}
\] & & 807
\(\mathbf{X 7 9}\) \\
\hline
\end{tabular}

\section*{BAND 3 T.V. CONVERTOR 183/196Mo/s.
 suecessful unit. comprising drilled ohassis, valves. coils, res. cond.,
ette, slightly modified version (Wireless World, \({ }^{2} \mathrm{ay}, 1954\) ), \(£ 2 / 5 /\)-post Iree Bend for blaeprint and wiring diagram. 1/6 post Iree Power Psok Componopks. inelading mains transf. and met.
foot., \(30 /\) oxtra. Provisiou has allowed on chassis for Bend 1-Band 3 switching. suitable for most types \\ }

RADID - GRAM GHASSIS 5 VALVE SUPERHET, LATEST 3 WAVEBANDS: \(-L . W . ~ 800 \mathrm{~m}-2000 \mathrm{~m}\), M.W. \(200 \mathrm{~m}-550 \mathrm{~m}\), SW. \(16 \mathrm{~m}-50 \mathrm{~m}\)
 dorizontal or vertical slation Names and 4 control knobs, wainut of lvory to choice. 4 position W/C switch, L.M.S. and Gram. P.U. sockets. Modern eirouitry, all ooils adjustable, dust corea and only quality com ponents nesed throughout. Delayed A.V.C. and neg. feedbaok. A.C. msins 200-250 v. Donble wound transi, isolates ohassi from mains. Aligned and ceimbrated ready or use.
BRAND NEW \& GUARANTEED \(£ 9.15 .0\) Carr. and ins, 4/6.
\(8^{\prime \prime}\) and \(10^{\prime \prime}\) speakers suitable for this chassis available.
7-Valve De Luxe, push-pull version, 7-watt output \(£ 12.10 .0\).

\section*{QUALITY FLUORESCENT FITTINGS}

Ideal for home or workshop. 4tt., \(220-250 \mathrm{v}_{\text {o, }}\) complete with tube, bsillast mit, etc. ready for use. Famous manufacturers' surplus offered at approx, hall price. Starter swittoh type 42/-. Quick start type 47/6. Carr. and ins., 7/6 extra

\section*{ELECTROLYTICS Leading Makes New Stock}


\section*{RESISTORS}

Carbon type. Pret, values 10 ohms10 megohms, \(20 \%\) Tol. \(\frac{1}{1}\) w. \(3 \mathrm{~d} . ; 1\) w. 5 d .

WIRE WOUND TYPES

\section*{Wire ends. Sllicone coated. 25 ohms-} 10,000 obms, 5 w ., \(1 / 3.10 \mathrm{w} .1 / 6.15 \mathrm{w}\)
I/Nie cord . 3860 ohms per It. 2a 100 ohma per tt." 2 way 6d. per ft., 3 way

\section*{LOUDSPEAKERS}
P.M. 3 OHM. 5in. Celes., \(17 / 6 ; 6 \mathrm{in}\). Celes. 18/6: \(7 \times 4 \mathrm{in}\). Goodman's Elliptioal. 18/8: 8in, Elac., 20/-; 8in. 25/-; 12in. Plessey, 35/-; 8ia. M.E. 2 k ohms field, tapped 0 O P trans. \(24 / 6\); 3 itin Elac \(17 / 6\)

\section*{S.T.C. RECTIFIERS}

EHT types K3/25 \(2 \mathrm{KV}, 4 / 3 ; \mathrm{K} 3 / 40\) \(3.2 \mathrm{KV} . .8 /-7 \mathrm{~K} 3 / 453.8 \mathrm{kV} . \mathrm{F} 6 / 6 ; \mathrm{K} 3 / 50\)

 120 mA ., \(5 / 9 ;\) RM4 250 甲. 275 mA .,


PRE-SET W/W POTS T.V. knurled slotted knob type. 25 ohms to 30,000 ohms., \(3 /-; 50.000\) oyms, \(4 / 7\) 50.000 ohms to 2 Megohms (carbon), 3/-
VOIUME CONTROLS Midget \(\log\) type, long spindies, all values 10,000 obms to 2 Magohms. Less sw., \(3 /\);-
 switch \(4 / \mathrm{l}\). Gnsranteed 12 months.

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THE MODEL CR50 BRIDGE measures from 10 pFd to 100 mFd and from I ohm to 10 megohms in fourteen ranges, having a otal scale length of over 120 inches. Indication of balance is given by a magic eye fed from a high gain amplifier valve. A leakage test is incorporated for condensers. The internal standards are Constantan \(1 \%\) resistors. Specially designed for bench use with case and panel of steel finished black crackle. Complete with all valves and instructions ready for immediate operation from \(200 / 250\) volt A.C. mains. Fully guaranteed and priced at only \(\mathrm{E} 7 / 18 /=\) plus \(4 / 6\) carriage and packing Hire Purchase \(44 / 6 / 3\) deposit and five monthly payments of \(17 / 3\).
SG50 SIGNAL GENERATOR covers \(100 \mathrm{kc} / \mathrm{s}\) to \(80 \mathrm{mc} / \mathrm{s}\) in six continuous ranges on fundamentals (NOT harmonics) either modulated 400 cps or CW. Uses EF91, 6C4 and RMI with double wound mains transformer. The scale is directly calibrated on all ranges with scale length over 60 inches. In de luxe olive green metal case with carrying handle and engraved matching scale of Perspex. The cost today is only \(£ 8 / 10 /=\) plus \(6 /\)-carriage and packing. Hire Purchase \(\mathbf{E 4 / 1 3 / -}\) deposit and 5 monthly payments of 18/7.

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Twin silver \(\frac{1}{2}\) amp.
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Up to \(100,000 \Omega\) Guaranteed \(\pm 3 \%\) on Turns \(\pm 5 \%\) on ohms.

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\section*{LORAN INDICATOR UNIT} and screen, but less valves and crystal. New condition, price 29/6.
PANORAMIC ADAPTORS. Visual display of signals on 3 in . C.R. Tube 3BPI. Band width \(200 \mathrm{Kc} / \mathrm{s}\). Suitable for receivers with IF's of \(450-475 \mathrm{Ke} / \mathrm{s}\), e.g., CRI 100 and
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RECEIVER CR100. Covers \(60 \mathrm{Kc} / \mathrm{s}\) to \(30 \mathrm{Mc} / \mathrm{s}\) in 6 ranges. 2 RF's and \(31 F\) 's, variable selectivity,
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MODULATION TRANSFORMERS Bendix type, for TA-12, push-pull 807's to plate and screen modulate two 807's in paralle!, 15/-. Collins type, 20 watts, 807 to 807, 12/6.
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8 Hen., \(100 \mathrm{~m} / \mathrm{mmp}, 8 / 6\); STC, potted, 8 Hen., \(100 \mathrm{~m} / \mathrm{amp}\), , \(8 / 6 ;\) STC, potred,
swinging approx. 5 Hen., \(300 \mathrm{~m} / \mathrm{amp}\)., \(12 / 6\); swinging approx. 5 Hen., \(300 \mathrm{~m} / \mathrm{amp}\)., \(12 / 6\);
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 shrouded, 8 Hen. \(250 \mathrm{~m} / \mathrm{mpp}\). ., \(200 \mathrm{~m} / \mathrm{mmp}\); 3 Hen., \(100 \mathrm{~m} / \mathrm{amp}\)., or 10 Hen., \(200 \mathrm{~m} / \mathrm{amp}\). (meg. feet slightly damaged) \(8 / 6\) each; (mtg. heet s.ightly damaged) \(8 / 6\) each
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MARCONI SIGNAL GENERATORS. Brand new, in original transit case, complete with spares, instruction manual and calibration charts. Type TF390G, 4-32 and \(50-100 \mathrm{Mc} / \mathrm{s}\). E 25 .
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AMPLIFIERS


MODEL MI-11220. Employs 2 6L6, 4 6J7, and 1 5U4. Output 12 watts at \(5-7.5-15-500\) ohms. For 190/250 v. A.C. mains. In grey crackled case, I7in. x lin. x 9in., we 3816 mains. In grey crackled case, \(17 \mathrm{in}\). . 1 lin. \(x 9\) in. Wt 381 b
Brand new and boxed. Price, less valves, \(£ 9 / 19 / 6\). Circuis Brand new and boxed. Price, less 5 val
supplied. Set of new boxed Vaives, \(59 / 6\).
APQ9. TRANSMITTER. Contains 931A electron \(\begin{array}{ll}\text { APQ9 TRANSMITTER. } \\ \text { multiplier (complete with resistor network, etc.), } & 2\end{array}\) 6AC7, I 6AG7, 2807 , and 2 blower cooled 8012 . Lecher Line tuning with rev. counter. Just the job for centimetric band experiments. Brand new, \(79 / 6\).
RT37/PPN2 BEACON TRANSMITTER-RECEIVER. \(214-234 \mathrm{Mc} / \mathrm{s}\). Slze \(13 \mathrm{in} . \times 10 \mathrm{in} . \times 5 \mathrm{in}\). Contains 5 3A5, \(214234 \mathrm{Mc/s}\). Slze 13 in . x Min. \(x\) Sin. Contains \(53 A \mathrm{~S}\),
3 IS5, I IR5, and 22 V . synchronous vibrators. Operaces IS5,
from 2
v. accumulator via 2 builr-in vibrapacks. Complete with telescopic mast antenna system (91/ft.) lightweight headphones, Technician Manual, super quality carrying haversack, cords, co-ax. cables, plugs, etc. Total wt. 28lb. BRAND NEW, boxed, American equipment, 72/6.

\section*{COMMAND RECEIVERS}

Complete with 6 valves: 3 12SK7, 1 12K8, I 12SR7, and I 12A6. Size \(5 \mathrm{Sin} . \times 5 \mathrm{tin} . \times 11 \mathrm{in}\). deep. In very good condtion. Less dynamotor. FREE CIRCUIT WITH EACH SET.
BC 455, 6-9 Me/s, 25/-. BC453, 190-550 Kc/s, \(59 / 6\). BC454, 3-6 Me/s, 27/6. BCXXX, 1.5-3 Mc/s. 65/-. BC454, NEW, BOXED, BC453 69/6. BCXXX, 75/-.
COMMAND TRANSMITTERS. Complete with 2
1625 ( 12 v. 807), I 1626, I 1629 and Crystal.
\(\begin{array}{llll}\text { BC457. } & \text { 4-5.3 Me/s (Xral } 4600 \mathrm{Ke} / \mathrm{s}), ~ 22 / 6 \text { New, } 29 / 6 . \\ \text { BCZZZ 2.I-3 Me/s (Xtal } 2500 \mathrm{Ke} / \mathrm{s} \text { ). }\end{array}\) BCZZZZ 2.1-3 Mc/s (Xtal \(2500 \mathrm{Ke} / \mathrm{s}\) ),

Unboxed, 29/6.
TWO-WAY MORSE TRAINING SETS, W/T Mk. 3. Consists of 2 valve oscillators (ARPI2's) (one with pitch control), for 1 or 2 operators. Has provision for creating 'atmospherics." In polished oak case \(12 \frac{1}{2} \mathrm{in}\). \(\times 10 \mathrm{in} . \times 8 \mathrm{in}\). wt. board, circuit and instructions, bur less batteries and phones
Ideal for Cadets, Scouts, etc. SNIP, 19/6.

\section*{METER BARGAINS}
\begin{tabular}{|c|c|c|c|c|}
\hline \multirow[t]{6}{*}{\begin{tabular}{l}
RANGE \\
50 Microamp. \\
100 Micruamp.
500 Microamp. \\
500-0.500 Miero \\
amp.
\end{tabular}} & TYP & SIzE & & \\
\hline & D.C. M/C & 2 fin . & Fluabh circ., scaled 0-100 & 59/8 \\
\hline & D.c. m/C & 2 tin. & Fluah otre., scaled 0-1,50 & \\
\hline & D.C. M/C & & Flush square; Weston & \\
\hline & ro- D.C. M/C & 2 j in. & Flush ciroular, scaled & \\
\hline & & & & \\
\hline Muliamp. & & 2 L & Flus & \\
\hline 10 Millimip. & D.C. M/C & \({ }_{2}^{2 \mu \text { in. }}\) & Dlush circular, blank acale & \\
\hline 100 Milliamp. & D.0. M1/C & 2 )1m. & Flush olrcular & 6 \\
\hline 130 Milliamp. & D.C. M/C & sin. & Fluah square & \\
\hline 200 Milliamp. & D.C. M/O & 2 j in. & Flush oircular & 10/6 \\
\hline 1 Amp. Th & Thermo-couple & 2)in. & Projecting circular & 8/9 \\
\hline 4 Amp. Th & Thermo-couple & 2 n . & Flush square & \\
\hline 30 Amp. & D.C. MC & 2in. & Projecting olreular & 8 \\
\hline 30-0-30 Amp. & D.C. M1 & 2 in . & Proj, elirc., car type & - \\
\hline 15 Volts & A.C. M/I & \({ }^{2} \mathrm{in}\) in. & Flush circular & 816 \\
\hline 300 Votrs & A.C. \(\mathrm{M} / \mathrm{I}\) & \({ }^{31 \mathrm{ln} .}\) & Flubh circular & 301- \\
\hline Volts & D.0. M/C & 2in. & Flush qquare & 10/9 \\
\hline
\end{tabular}

METER RECTIFIERS. Salford Instruments, \(1 \mathrm{~mA}, 8 / 6\); \(5 \mathrm{~mA}, 6 / 9\); S.T.C., 2 mA., as used in E.M.I. Output Meter, \(5 / 6\). All are full wave bridge and brand new.
MINIATURE MODEL MOTORS, 2 in . long \(\times \operatorname{la}_{\frac{3}{4} i n}\) diam \(\frac{3}{16} \mathrm{i}\). diam. spindle, we. \(50 z\). Will work from 6 y . dry battery and are reversible. Has ball bearings. 8/6.

INDICATING UNIT 277. Case size \(5 \frac{1}{2} \mathrm{in} . \times 7 \mathrm{in} . \times 12 \mathrm{in}\). deep. Contains lin. C.R. tube type VCR522 (same as that used in G.E.C. "Miniscope"), 4 VR91, 2 VR92, and a host of useful, modern components. Fitsed with "Focus" and "Brigheness" controls, etc. Should convert to useful miniature oseilloseope. All tubes tested. 59/6.
ART7 MAINS TRANSFORMERS. Fully shrouded, drop-through, \(3 \frac{\mathrm{in}}{\mathrm{in} .} \mathrm{x} 3 \mathrm{i} \mathrm{in}\). Primary \(110 / 125 / 150 / 210 / 240 \mathrm{v} .50-60 \mathrm{c} / \mathrm{s}\). Secs., \(325-0-325 \mathrm{v} .100 \mathrm{~m} / \mathrm{amp} .{ }^{6} 6.3\) v., 3.5 amp., and 5 v. 2 amp. BRAND NEW exUS.A. 2916
REPLACEMENT. MAINS TRANS. for R C.A. Amplifiers, Pri. 105/115/125 Y., 50-60 \(\mathrm{c} / \mathrm{s} .\), Secs., \(380-350-0-350-380\) v., \(170 \mathrm{~m} / \mathrm{amp}\). \(6.3 \mathrm{v} ., 3 \mathrm{amp}\)., C.T. and \(5 \mathrm{v} ., 3 \mathrm{amp} .4 \mathrm{in}\). \(43 \mathrm{in} . \times 43 \mathrm{in}\). high. Fully shrouded, \(16 / 6\). TRANSFORMER BARGAINS. Brand new ex-manufacturer's surplus drop-through. Primary 200/250 volts 50 cps. Secondary \(310-0-310 \mathrm{v} .70 \mathrm{~mA} ., 6.3 \mathrm{v}\) @ @ \(3 \mathrm{a}, 4 \mathrm{v}\). @ 2 a . Can be used with either 4 v . or 6.3 v . rectifier. Only \(9 / 6\) plus \(1 / 6\) post. A similar type transformer 325-0-325 \(100 \mathrm{~mA} ., 6.3 \mathrm{v} .4\) a., 4 v . 3 a. only \(12 / 6\).
INSTRUMENT TRANSFORMERS, Parmeko. 230 v. A.C. input. 0-65-130-195 v. \(85 \mathrm{~m} / \mathrm{Amp} .6 .3 \mathrm{v} .5 \mathrm{amp}\). and 6.3 v .3 Amp . Shrouded, \(3 \frac{1}{4} \mathrm{in}, \times 3 \frac{3}{4} \mathrm{in}, \times 3 \frac{3}{4} \mathrm{in}\). hizh. \(15 /\)-. HEAVY DUTY LT TRANSFORMERS. 230
\(5-0-5\).
v., and mains input. Socs. 5-0 \(5-0-5\) v., all at 5 amps, each \(5-0-5 \mathrm{v}^{2}\), and \(5-0-5 \mathrm{v}\), al at 5 amps, each
winding. \(5,10,15,20,25\), or 30 v . at 5 amps., or other possible combinations. \(4 \frac{1}{2}\) in. \(x\) 44 in . x 6in. high. Wct. 12 lb . BRAND NEW, 29/6. ANOTHER, 230 v . A C. mains input, Two separate secondary windings, each 14 v . С.T. \(12 \mathrm{amps} .7,14,21\), or 28 v . at 12 amps., or other possible combinations. Wt. 24 Ib . Ex-Admiralty. Brand new, 42/6. G.E.C. \(200-250 \mathrm{v}\). A.C. mains input. 30 v . (tapped at 10 v.) 36 amps output. \(5 \frac{1}{2} \mathrm{in} . x\) 6in x 7 in , high. We. 24lb., 55/-.
SELENIUM BRIDGETIFIERS. Funnel cooled. A.C. input 45 v. R.M.S. Funnel cooled. A.C. input 45 v . R.M.S.
D. C output \(30 \mathrm{v} 10 \mathrm{amp} .47 /\).6 each. HEAVY DUTY SLIDER RESISTORS. 250 watts. Rated to carry 25 amp . . 4 ohm resistance. For charging boards, etc. Worm drive. On metal stand 9 in . \(\times 4 \mathrm{in}, \times 6 \mathrm{in}\). high. BRAND NEW \(7 / 6\) each. ANOTHER, 12 amps, 1 ohm, 150 watts, \(6 / 6\). ALSO 14 ohms, graded 1 to 4 amps., \(7 / 6\).
RESISTORS.
RESISTORS. Latest miniature insulated Dubilier \(\frac{1}{2}\) watt type B.T.S. Wire ends.
Useful values. ONLY \(10 /\) for 100 assorted! Erie, etc. \(\frac{1}{2}\) and + watt, 1 gross assorted \(10 /\) R.F. UNITS. ALL BRAND NEW and BOXED. RF24, 20-30 Mc/s, 12/6. RF25 \(40-50 \mathrm{Mc} / \mathrm{s}\). , \(17 / 6\). RF26, \(50-65 \mathrm{Mc} / \mathrm{s}\)., 29/6 and RF27, 65-85 Mc/s.; 32/6.
HIGH VOLTAGE POWER PACKS ex-U.S.A. Brand new Input 115 voles A.C., ox-put 1,000 volts at 250 mA . Complete with 21616 valves and I spare. Double choke and paper smoothing. ES/5/-. RY POWER ONITS. in grey metal case. Input 12 v . D. C. Output 180 v . at \(60 \mathrm{~m} / \mathrm{a}\). Complete with all smoothing and filtering. Size \(12 i n,{ }^{2}\)
Sin. We. I61b. BRAND NEW, \(19 / 6\).
BLOCK PAPER CAPACITORS, T.C.C. \(8 \mathrm{mFd} ., 1 \mathrm{kV}\). Wkg., upright mag., size 4 in . \(x\) 2tin. \(x 5\) in. high. Brand new and boxed. TWO for 151.
THREE-CORED CABLE. 23/36, rubber ins. circular, padded, cotton covered, maroon. 12 yds., \(9 /-\) or or \(100 \mathrm{yds.,5} 59 / 6\). METAL RECTIFIERS, \(250 \mathrm{v}, 100 \mathrm{~m} / \mathrm{amps}\). \(6 / 9 ; 230 \mathrm{v} .60 \mathrm{~m} / \mathrm{amps} ., 5 /-\). Many other types in stock.
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\(8 \mathrm{mfd} .450 \mathrm{v}, 2 / 6\) each; \(16 \mathrm{mfd} .450 \mathrm{v} ., 3 / \cdot 132 \mathrm{mfd} .450 \mathrm{v.} 4 /,-; 8 \times 8 \mathrm{mfd}\).
\(450 \mathrm{v}, 39 ; 8 \times 16 \mathrm{mfd} .450 \mathrm{v}, \mathrm{e} 41-; 16 \times 16 \mathrm{mid} .450 \mathrm{v}, 4 / 6 ; 32 \times 32 \mathrm{mfd}\).
350 v., \(5 /\)-. Bias Condensers: \(25 \mathrm{mfd} .25 \mathrm{v} ., 1 / 6 ; 50 \mathrm{mfd} .50 \mathrm{v} ., 1 / 9\).
Please note we can offer special discounts for quantities.
ELECTROLYTIC CONDENSERS. Manufacturers' Surplus, in perfeat
\(16 \mathrm{mfd} .375 \mathrm{v}, 2 /-; 24 \mathrm{mfd} .350 \mathrm{v} ., 1 / 6 ; 24 \mathrm{mfd}, 450 \mathrm{v},. 2 / 3\).
BIAS CONDENSERS: \(1,000 \mathrm{mfd} .12 \mathrm{v}, 1 / 6 ; 25 \mathrm{mfd} .25 \mathrm{v} ., 1 / 3 ; 50 \mathrm{mfd}\). 12 v , \(1 / \mathrm{m}\).
BLOCK PAPER CONDENSERS. \(12 \mathrm{mfd} 250 \mathrm{v} ., 7 / 6 ; 8 \mathrm{mfd} .600 \mathrm{v}\)., 7/6; 4 mfd . \(400 \mathrm{v} ., 3 / 6\). We carry a large stock of block paper type condensers. We invite your enquiries.
MIDGET MICA CONDENSERS. .0001, .0002, .0003, .0004, .0005, 5/- per dozen.
200 Assorted Moulded Mica Condensers, popular values.......... \(£ 210\). 0 200 Assorted Silver Mica Condensers, popular values................ \(£ 2\) 10 0 200 Assorted Carbon Resistors, \(\downarrow, \frac{1}{2}\) and 1 watt. Good selection el 100 PAXCOLIN SHEET. \(18 \vee 4 \frac{1}{2} \times \frac{1}{10} \mathrm{in}, 1 / 6 ; 10 \times 10 \times \frac{1}{3} \mathrm{in}\). \(1 / 6 ; 20 \times 10 \times \frac{1}{13} \mathrm{in}\) \(3 /-; 10 \times 10 \times \frac{1}{16} \mathrm{in} ., 2 /-; 20 \times 10 \times \frac{1}{16} \mathrm{in}^{2}, 4 /\). Minimum P' \& Pkg. \(1 / 6\).

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" S " Meters as used in S .27 receivers, \(\in 2 / 10 /=00-200 \mathrm{~m} / \mathrm{a}\)., \(2 \frac{1}{2} \mathrm{in}\)., \(10 / 6 ; 0-300 \mathrm{~m} / \mathrm{a} ., 2 \mathrm{t} \mathrm{in} ., 10 / 6 ; 0-500 \mathrm{~m} / \mathrm{a}, 2 \frac{1}{2} \mathrm{in}\)., \(12 / 6 ; 0-40 \mathrm{~m} / \mathrm{a}\). (scaled \(0-3 \mathrm{kV}.), 10 / 6: 0-100 \mathrm{~m} / \mathrm{a}, 2 \mathrm{in}\). (scaled \(0-300^{\prime \prime}\) 'mag feed"), \(7 / 6\).
B.T H. CRYSTAL DIODES, I/3. Very special price for large quantities. RESISTORS. \(\frac{1}{4}\) watt, \(2 / 6 \mathrm{doz}\); \(\frac{1}{2}\) watt, \(3 / \cdot \mathrm{doz}\); I watt, 4/- doz; 2 watt, 6/- doz.
W.W. RESISTORS. 5 waft, \(1 / 6\); 10 watt, 2/6; \(15 \mathrm{watt}, 3 /-; 20\) watt, 3/6. We carry stocks of resistors from 2 watt to 150 watt W.W. Your enquiries invited.
HIGH STABILITY RESISTORS. \& watt \(5 \%\), \(6 \mathrm{~d} .0 \frac{1}{2}\) watt \(5 \%, 9 \mathrm{~d} .:\) I watt \(5 \%, 1 /\). A few values in \(1 \%\) and \(2 \%\) still available.
ALL ORDERS FOR RESISTORS C.O.D. PLEASE, AS WE CANNOT GUAR. ANTEE TO STOCK ALL VALUES.
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SEMI-MIDGET 2-GANG. . 0005 Condenser, size \(21 \times 2 \times 1 \frac{1}{4} \mathrm{in} ., 6 / 2\) each.
PUSH-BUTTON UNITS. 10 -way, complete with escutcheon and knobs, \(10 / 6\) each; 4 -way, \(2 / 6\) each; knobs \(3 /\) - doz extra.
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\hline FERRANTI & 65 & 0 \\
\hline STARK 70 & 48 & 0 \\
\hline TAYLOR 70 & \(\pm 9\) & \\
\hline
\end{tabular}
\begin{tabular}{|c|c|c|c|}
\hline AIRMEC & £45 & 0 & 0 \\
\hline AVO & ¢25 & 0 & 0 \\
\hline MARCONI & 625 & 0 & 0 \\
\hline RCA & ¢18 & 0 & 0 \\
\hline TAYLOR I70A & \(\underline{20}\) & 0 & 0 \\
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\hline 1 Milliamp & 2 in . & MC/FR & 17/6 \\
\hline 1 " & 2 in , & MC/FS Elliott 50/87 & 27/6 \\
\hline 5 & 2 in . & MC/FS & 10/6 \\
\hline 30 & 21 in . & MC/FR & 12/6 \\
\hline 100 & 2 lin . & MC/FR & 12/6 \\
\hline 200 & 21 in . & MC/FR & 12/6 \\
\hline 500 & 2 in . & MC/FR & 12/6 \\
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Degree or similar qualification necessary. Candidates who have been associated with production will receive preference.

Duties include the design of domestic radio and television, high fidelity audio equipment and miniature transmitter receivers.

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For development of (1) VHF and UHF receiving aerials. (2) Interference suppressors and screened compartments. (3) Circuit protection. devices (fuses, cutouts, etc.). Minimum qualifications Grad.I.E.E., H.N.C. (advanced) or equivalent.

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To prepare clear and concise reports from laboratory notes and précis suitable for advertisements, trade and internal circulation. This is a position for a man or woman with moderate technical qualifications, but a sound general education and a flair for writing are essential.

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Ditton Works, Cambridge
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with experience in the field of
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Apply in writing giving details of qualifications and experience to the PERSONNEL MANAGER Quoting Reference M.D.


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This is a four year course and involves alter nating periods of 6 months' full time studying and interesting varied practical work as these Laboratories. It is designed primarily for Physicists and Electrical Engineers and will lead to the London 8.Sc. (Special Physics) or B.Sc. (Eng.) Applications are also invited from potential chemists, metallurgists ared also invited from potential chemists, well as from those interested in a career in patents work. Apart from a generous salary all College and examination fees will be paid.
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Engineers are required for Senior and Junior positions in the Television and Radio Development Departments of a well known Manufacturer in the West London area.

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Please write fully, in confidence, stating age, experience, etc. to Box No. 4256.

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Well-known manufacturers of electronic equipment are willing to manufacture and place on the market:-

\section*{THAT NEW IDEA DF YOURS}
(Our own staff is aware of this advertisement). Royalty will be paid on all sales. For further details write Wireless World Box No. 4236.

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TRANSISTOR CIRCUIT ENGINEERS for development of circuitry around transistors.
THERMIONIC VALVE ENGINEERS for development and production engineering of broadcast and special valves.
MECHANICAL ENGINEERS for development of devices for the mechanisation and control of valves and cathode ray tube manufacturing processes.
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The Cable. Telecommunication and Electronic Industries in which the Companies are principally engaged, employ advanced techniques of high voltage, electronic and semi-conductor engineering. As the Main Works are in London the technical staff can participate in the activities of the professional institutions and attend a wide range of lectures and courses. There is a liberal pension scheme in operation and assistance with house purchase, if required, will be available to successful candidates.
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Write stating the nature of the work in which you are interested, giving full particulars of training, qualifications and experience, to the Technical Director (Group F), Siemens Brothers \& Co. Limited, Woolwich, S.E.18., or to the Director of Engineering (Group F), Edison Swan Electric Co. Ltd., Brimsdown, Middx.


EDISWAN

SIEMENS BROTHERS \& CO. LIMITED EDISON SWAN ELECTRIC CO. LIMITED

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RCA Model AVT-22-B 5 KW COMMUNICATIONS AND BROADCAST TRANSMITTER (for export only)
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* We are pleased to offer the RCA AVT-22-B Transmitters and the associated equipment (Master Oscillator, Power Transformer, Modulation Transformer, Crystal Oscillator, Electronic Keyer, 3-phase Auto Transformer (400/230 v. 18 KVA) and Portable Broadcast Amplifier).
The transmitter combines the highest efficiency compatible with communications and broadcast standards. Design features include drop-down and slide-out units for easy access, protective interlocks, optional electronic telegraph keying and safe operation in the temperature range \(-40^{\circ} \mathrm{C}\) to \(+55^{\circ} \mathrm{C}, 95 \%\) humidity.
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Power Outpue: 5 kW .
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EMIS5IONS: AI, A2 and A3
MODULATION: High Level Class B.
Frequency response of the broadcast version: 30 c.p.s. to \(15 \mathrm{Ke} / \mathrm{s}\).
SPECIAL FEATURES: On telegraph, electronic keying up to 400 words per minure. Facllity for optional use on half power.
The equipment matches a resistive antenna or line load 50 to 1,000 ohms.

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The AVT-22 Master Oscillator is a highly stable source of RF power which may be utilised as an exciter at any frequency between 2 and \(20 \mathrm{Mc} / \mathrm{s}\). In conjunction with the AVT-22-8 the unit is used for frequency modulation of telegraphically the unit
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Oscillator Unit
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- TR-50 SERIES We are now in a position to speed up delivery of our well-known 5eries of TR-50 Transmitter-Receiver Installation ( 50 watt power output, 1.5 to \(12 \mathrm{mc} / \mathrm{s}\), 12 v . 24 v . or \(115 / 230 \mathrm{v}\). A.C., full set of equipment spares for 5 years' operation).
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Readers are warned that Government surplus components and valves which may be offered for sale through our displayed or classified columns carry no manufacturers' guarantee. Many of these items will have been designed for special purposes making them unsuitable for civilian use, or may have deteriorated as a result of the conditions under which they hove been stored. We cannot undertake to deal with any complaints regarding any such items purchosed.

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IDDYSTONE 740 receiver, perfect 'phones, speaker, any trial; offers.-Tel. Elou. 2933. A RMSTRONG RF41 chassis. minti \(165 / 29\) Abercorn Place. N.w.8. Mai. 5208. HRO Rx's and colls in stock, also AR88. 0 T I. Service 254, Grove Green Re London, E.11. Ley. 4986. [0053

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[^4]:    $\dagger$ Patent applied for.

[^5]:    * Electronics Engineering Dept., British Thomson-Houston Comspany.
    $\dagger$ "Stereophonic Sound." J. Moir, March, 1951. "TwoChannel Stereophonic Systems,"* F H. Brittain and D. M. Leakey, May \& July, 1956.

[^6]:    *In the July to September issues inclusive.

[^7]:    $\dagger$ "Transistor Electronics," by A. W. Lo and others. (PrenticeHall, Inc.)

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