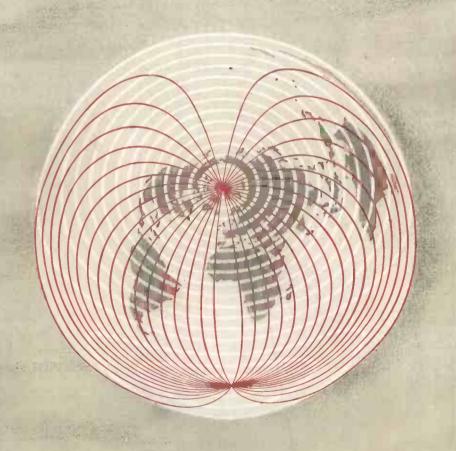
AUGUST 1955 TWO SHILLINGS

Wireless World

Radio · Electronics · Television





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AUGUST, 1955

Wireless World

RADIO, ELECTRONICS, TELEVISION

Managing Editor: HUGH S. POCOCK, M.I.E.E.

Editor:
H. F. SMITH

AUGUST 1955

In This Issue

VOLUME 61 NO. 8 PRICE: TWO SHILLINGS

FORTY-FIFTH YEAR
OF PUBLICATION

Editorial Comment	353							
Colour Television Standards	355							
Short-wave Conditions	355							
World of Wireless ,	356							
Inexpensive Wave Analyser. By M. G. Scroggie								
Letters to the Editor	367							
Band II F.M. Tuner Unit. By L. Hampson								
New Navigational Aids	3 75							
The Automatic Factory	378							
Wide Range Electrostatic Loudspeakers—3. By P. J. Walker								
High-Fidelity Home. By Richard Arbib	385							
Transistor Equivalent Circuits—2. By W. T. Cocking								
Transmitting Colour Information	393							
The Cascode. By "Cathode Ray"	397							
Electronic Teleprinting	401							
Random Radiations. By "Diallist"	402							
Unbiased. By "Free Grid"	404							

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32. HIGH GAIN VIDEO AMPLIFIER

The gain of a video amplifier with a given bandwidth, using a high slope r.f. pentode, is limited by the output or load capacitance. A typical cathode compensated circuit using an EF80, with a bandwidth of 3Mc/s (6dB), a maximum video output voltage of 50V, and a load capacitance of 26pF, has a gain of 11. The gain can be increased and the bandwidth maintained if the effective load capacitance is reduced by means of a cathode follower. The practical circuit shown in Fig. 1 has a gain of 22. A Mullard PCF80 triode pentode combines the functions of video amplifier and cathode follower in one envelope.

FIG I 0-8mA 3-7mA SI5kΩ IW VIDEO 3·5V

Video Amplifier

The cathode of the pentode (Fig. 1) is held at about + 3.5V by the pentode cathode current and the bleed current through the 27kΩ resistor. With the screen grid at 190V this gives an anode current of 3.5mA. The permissible anode current swing is limited by the maximum screen grid dissipation rating of 0.5W, therefore the output voltage is limited to about 80V. The dynamic characteristic of a typical PCF80 pentode shows that the corresponding maximum limit of the anode current is 8.8mA. effective slope is 1.5mA/V, therefore the gain gmR1 = $1.5 \text{mA/V} \times 15 \text{k}\Omega = 22.5.$

Cathode Follower

Mullard

The total capacitive load of a video stage is made up, in a typical good design, approximately as follows: Video output 4pF, C.R.T. 8pF, Synchronising separator 6pF,

Noise suppressor 2pF, Strays 6pF = 26pF.

If the total gain of the stage is to be greater than 20 and a cathode follower output (with a gain less than 1) is to be used, the amplifier must have a higher gain—say 25. To achieve this the output capacitance must be less than 11pF. With a cathode follower the capacitive load of the amplifier is reduced to: Video output 4pF, Cathode follower input 3pF, Strays 3pF = 10pF, which is within the 11pF limit. The overall response is limited by the cathode follower, which must be capable of developing an output voltage of at least 50V across a capacitive load of 26pF.

> Further notes on the circuit are available with reprints of this advertisement. Reprints of all advertisements in the Valves, Tubes and Circuits series may be obtained free from the address below.

Cathode Current Requirements

The cathode current has two components: a current which maintains the signal across the cathode resistor, and a transient current which charges the output capacitance. The valve must supply both currents. If it failed to do so the cathode voltage would not follow the grid voltage. If the cathode cannot rise at the same rate as the input the valve will run into grid current, and the positive going edge of the signal will be distorted. Alternatively, if the cathode cannot fall at the same rate as the input, the valve will be cut off and the negative going edge will be distorted. In general the second alternative is worse, as cut-off is an absolute limit to change in the cathode current. The onset of grid current does not preclude further change if the source impedance of the driver stage is reasonably low. As each input pulse has two transients associated with it the total current swing is the maintenance current plus twice the transient current. To prevent cut-off the standing cathode current must exceed the sum of the transient and maintenance currents.

FIG. 2

Derivative Correction

It is possible to compensate for a poor fall time in a cathode follower by adding to the output a voltage proportional to the derivative of the video input signal. This voltage, which can be obtained by the inclusion of a resistor in the triode anode circuit, may be a.c. coupled to the c.r.t. grid when the cathode follower is driving the c.r.t. cathode. The derivative correcting voltage is thus added in the correct phase. It is found that this arrangement increases the contrast of the high definition bars of test card 'C' and improves the vertical edges; but it tends to produce ringing and 2.5Mc/s oscillation, therefore the use of derivative correction with this circuit is not recommended.

Performance

In a 21-inch tube receiver the recommended circuit (Fig. 1) gives a composite video output up to 80V before there is any visible sign of cathode follower overload. The frequency response, measured with a sine wave input and a 50V peak-to-peak output, is shown in Fig. 2. The stage gain is 22.

Wireless World

AUGUST 1955

VOL. 61 No. 8

Tape Bookmark

HE rapid growth of tape recording for domestic and business purposes is largely due to the excellence of what is rather pompously called the tape transporting system. Mechanisms for winding the tape from one spool to the other are, generally speaking, remarkably effective; the clumsiest of us find little difficulty in operating even the cheaper and less

refined machines.

Judging by letters from readers, however, the domestic tape recorder is widely thought to be deficient in one respect: it is difficult to "find the place" on the tape at which any particular item starts. True, some kind of rough-and-ready gauge is often provided, but it does not give precise location. Most home recordists make use of scraps of paper inserted, like a bookmark, between adjacent turns of tape. This method, though fairly adequate when the "fast wind" mechanism can be smoothly and accurately controlled, is often considered to be crude. What seems to be wanted by a wide circle of users is a device which, by mechanical, magnetic, electrical, or even electronic means, will indicate precisely the part of the tape required. Something much more precise than the classical bookmark is in fact called

A device that went a long way towards meeting these needs was described in Wireless World for April, 1955. This embodied a selector switch actuated by pads glued to the tape at appropriate positions. Another method, used in certain machines, calls for the use of a revolution counter.

Perhaps the most novel of the various place-finding suggestions put forward by readers is for a system of signals imposed on the tape and audible only during the "fast wind"—forward or back. The idea is that these pulse signals, which could be arranged according to a code, would be recorded at a very low frequency and so would not be heard (or at least would not be obtrusive) at normal playback running speed. This method has obvious attractions—and, probably, what might be considered equally obvious limitations and drawbacks. For one thing, it would not be readily applicable to existing machines in which the playback head is inoperative during "fast wind." However, the

method seems worth investigating—always assuming, of course, that no satisfactory alternative that is inherently simpler and cheaper can be devised.

Colour Television Experiments

100 much significance need not be attached to the standards (set out on another page) chosen for the B.B.C. experimental transmissions in colour television, to begin in the autumn. It was more or less a foregone conclusion that a start would be made with a version based as closely as possible on the standards devised by the National Television Stan-The so-called dards Committee of the U.S.A. "British N.T.S.C." system involves very little change or addition to the B.B.C.'s existing monochrome transmitters. Another factor which might well have influenced the decision is that there is now a large amount of American experience with this system on which to draw. Further, there can be no doubt that the technical elegance of the system is attractive, except perhaps for the complexity of the receivers that go with it. Presumably the main purpose of the experimental transmissions is to allow that section of industry concerned with receiver design and manufacture to carry out field tests and to gain practical experience of colour television.

So far there has been no mention of experimental transmissions using other systems. No doubt, however, these will come; there is no need to make an over-hasty decision as to what system is finally to be adopted in this country. Indeed, a regular colour service is not expected for a long time.

In considering these questions, it must be remembered that all factors affecting British television must finally be decided by the Postmaster-General, who, in his turn, is advised by the Television Advisory Committee. The T.A.C. has uncompromisingly recommended a compatible system, and the British N.T.S.C. standards of the forthcoming transmissions satisfy that requirement. However, there may well be a change of view on this subject before the time comes for making a final decision.

Growth of Aircraft Radio

IN the Golden Jubilee celebrations of the Royal Aircraft Establishment at Farnborough this year, a substantial proportion of the exhibits were concerned with radio aids to navigation, communication and the application of radio and electronic techniques to the guidance of rocket missiles and the simulation

and computation of their performance.

An historical exhibit, arranged in chronological order, began with the Sterling Type 52A spark transmitter and the Model TF valve receiver (detector and 2 l.f. stages) of the 1914-18 war, and led up to the multi-channel crystal-controlled equipment in use to-day. One was reminded of the fact that although the possibilities of flying and of wireless communication were both realized at the turn of the century, a decade was to pass before they could come together. With the aerofoils and engine powers then available no designer, harassed as he was by thoughts of how best to save weight, could afford a second glance at the ship-and-shore type of equipment which was the stock-in-trade of the then infant radio industry.

There is still constant pressure from commercial and military aircraft designers to reduce the weight and size of radio equipment to make room for paying passengers or more armament, and this is being met by increasing miniaturization, which was adequately

represented in the exhibition.

The current work of the Radio Department at R.A.E. covers a wide field including the development of sono-buoys (in which the noises of submarines are picked up by hydrophones and relayed to a searching

aircraft by radio) and the detailed investigation of the problems of installing aerials with reasonably omni-directional characteristics on high-speed aircraft.

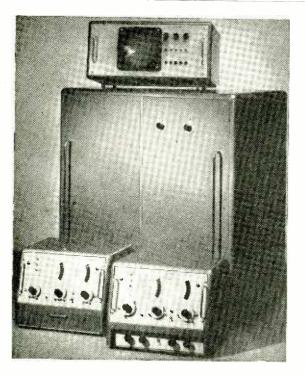
Electronic methods, once a useful alternative to established methods of physical measurement, can now be said to dominate all branches of aeronautical research. They reach their zenith at Farnborough in the "Tridac" analogue computor for guided missile problems, which occupies the whole of a special building and calls for primary power of the order of hundreds of kilowatts for the functioning of its many circuit elements.

Disc Recording and Reproducing Characteristics

IT has often been said that the ideal disc recording or reproducing characteristic should be one which is also easily realizable with simple circuitry. This provision is met in the proposals contained in the revised British Standard 1928:1955 which gives curves for fine-groove and coarse-groove recordings, and formulæ for their derivation in terms of the time-constants of simple R-C networks. These recommendations are based on C.C.I.R. standards and take into account the recommendations provisionally agreed by the International Electrotechnical Commission at their Philadelphia meeting last year.

The revised standard, which is obtainable, price 6s, from the British Standards Institution, 2, Park Street, London, W.1, includes specifications of commercial and transcription disc dimensions, stylus tip radii and concludes with a discussion of the arguments for standardization of the recording and/or the repro-

ducing characteristic.



Radar Simulator

EQUIPMENT for the training of radar operators and for the synthesis of tactical air exercises without the use of aircraft has been designed and developed by C. E. G. Bailey and J. Somerset Murray in association with the Solatron Electronic Group, Ltd., Thames Ditton, Surrey.

Aircraft are represented by control units which feed signals (analogous to speed, rate of climb or dive, rate of turn and direction of flight) into a computor unit. This unit integrates the factors governing range, bearing and height and translates the result into pulses for transmission to one or more radar display units.

Extraneous effects such as tropospheric refraction can be taken into account, and there are facilities for simulating jamming, either of the reflecting type ("window") or of the active noise-generating type.

The essence of the design is flexibility so that future as well as current characteristics of aircraft performance and radar systems can be simulated.

Typical assembly of Solatron radar simulator units. The main cabinet houses the computors. A display unit is shown on top. In the foreground are two aircraft control units.

COLOUR TELEVISION STANDARDS

For Forthcoming B.B.C.

Test Transmissions

XPERIMENTAL colour television transmissions based on the American N.T.S.C. compatible system will be made by the B.B.C. from Alexandra Palace this autumn as part of the general investigations into the best type of colour system for this country. A specification of the standards to be used shows the method of transmission to be a scaled-down version of the N.T.S.C. system, with the colour information transmitted by means of a sub-carrier within the existing 6.75-Mc/s monochrome channel.

An article on page 393 of this issue elaborates on the principles of the "British N.T.S.C." system, so it is unnecessary to add here more than the bare facts of the B.B.C. specification. First of all, the existing black-and-white transmission from Alexandra Palace will remain as it is and form the "luminance" or brightness component of the complete colour signal. Simultaneously a "chrominance" or colouring signal will be transmitted in the form of two sets of a.m. sidebands of two suppressed carriers in quadrature, these having the common frequency of 525/2 times the line scanning frequency relative to the 45-Mc/s picture carrier—in fact 2.6578125 Mc/s. These two "chrominance" components, known as the E₁ and E_Q signals (see article) will carry respectively wideband colour information up to 1 Mc/s and narrowband colour information up to 340 kc/s.

The "chrominance" or colouring sync signal will consist of a reference burst of 9 cycles of sub-carrier frequency transmitted during the "back porch" black-level period following each line sync pulse. It will not occur during the eight broad pulses of the frame sync period. Details can be seen from Fig. 1, which shows that the sync burst penetrates into the picture region and that the "chrominance" signal can do the same and also rise above peak white.

The complete colour signal, in terms of the total

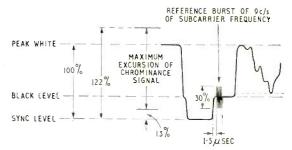


Fig. 1. Video waveform showing the relation of the added chrominance signal and sync burst to the existing monochrome signal.

video voltage applied to the transmitter modulator, is composed as follows:—

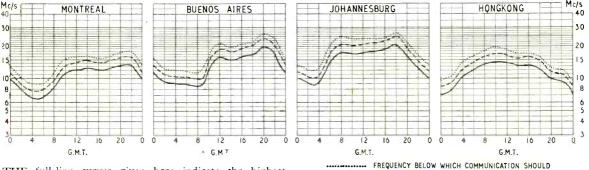
 $E_{\text{M}}=E_{\text{Y}}+K\left\{ E_{\text{Q}}\sin(\omega t+33^{\circ})+E_{\text{I}}\cos(\omega t+33^{\circ})\right\}$ Here E_{Y} (the "luminance" or black-and-white signal) is made up of $0.3E_{\text{R}}+0.59E_{\text{G}}+0.11E_{\text{B}}$ while $E_{\text{Q}}=0.41(E_{\text{B}}-E_{\text{Y}})+0.48(E_{\text{R}}-E_{\text{Y}})$ and $E_{\text{I}}=-0.27(E_{\text{B}}-E_{\text{Y}})+0.74(E_{\text{R}}-E_{\text{Y}}).$ The angular frequency ω is 2π times the frequency of the "chrominance" subcarrier, while the phase reference is the phase of the sync burst plus $180^{\circ}.$ The factor K indicates that various ratios of "chrominance" to "luminance" between 1.0 and 0.3 may be used in certain experiments.

Comparative tests may be carried out with the system locked to and unlocked from the 50-c/s mains. With unlocked operation the sub-carrier frequency will be the 2.6578125 Mc/s mentioned above, but in the locked condition it, and the frequency difference between the vision and sound carriers, will change directly with the mains frequency.

The experimental transmissions will, of course, be made outside normal programme hours, and readers who manage to receive them will perhaps get a foretaste of the effect of compatible colour—detrimental or otherwise—on the picture quality we are used to at present.

SHORT-WAVE CONDITIONS

Predictions for August



THE full-line curves given here indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during August.

paths from this country during August.

Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.

BE POSSIBLE FOR 25% OF THE TOTAL TIME

--- --- PREDICTED AVERAGE MAXIMUM USABLE FREQUENCY

FREQUENCY BELOW WHICH COMMUNICATION SHOULD

BE POSSIBLE ON ALL UNDISTURBED DAYS

WORLD OF WIRELESS

Show News + Anglo-French TV Link +

New Television Stations

Earls Court

NEARLY 50 per cent of the 90 or so manufacturers who will be exhibiting at this year's National Radio Show are set makers; the others being makers of components and accessories. The remaining 30 exhibitors are either users of radio—such as the B.B.C., the Services and Government Departments—wholesalers, societies and associations, and organizations providing services for the industry.

The show opens at Earls Court, London, S.W.5, on August 24th for ten days, with a preview for overseas

and invited guests on the 23rd.

In addition to the facilities provided on the exhibitors' stands for the demonstration of Bands I and III television sets, there will again be a display of some 100 receivers in Television Avenue. These receivers will be tuned to Band I. Incidentally, as in previous years, the Band I carriers piped round the show will be in Channel 4 to avoid interference from the London transmitter. Channel 8 is being used for the Band III demonstrations.

Each day at 2.30—except on Saturdays—a discussion meeting is being arranged by the British Radio Equipment Manufacturers' Association to bring together retailers' servicemen and representatives of the industry to discuss the servicing of f.m. receivers and the problems of f.m. aerial installation. Tickets must be obtained from B.R.E.M.A., 59, Russell Square,

London, W.C.1.

Cross-Channel TV Link

THE first section of the permanent Anglo-French television link, ordered by the B.B.C. from the Post Office last January, will be completed in September. This section consists of a two-tube co-axial cable between London and St. Margaret's Bay, Kent. The next section, consisting of a two-way cross-Channel radio link, will not be completed for three years.

However, so that we can participate in international programme exchanges (the next is planned for this autumn) the cables will be extended temporarily from St. Margaret's Bay to Swingate, near Dover—a distance of about two miles—where a temporary cross-

Channel radio station is set up.

It is understood that the cost of laying the cable, which is permanently rented by the B.B.C., and the cost of building the terminal station at St. Margaret's Bay will be nearly £750,000.

NATIONAL RADIO EXHIBITION

WIRELESS WORLD SHOW NUMBERS

September: Show Guide. Plan of the stands at Earls Court, with stand-to-stand guide to the exhibits.

October: Show Review. An analysis of design trends in television and sound broadcast receivers.

TEMPORARY 200-ft mast carrying the 8-stack array for the Croydon I.T.A. transmitter. Marconi's are supplying the complete installation for the temporary station due to open on September 22nd. The mast was completed early in July and the sound transmitter is installed.



B.B.C. Television Progress

DURING the past few days the permanent television station at Divis, near Belfast, has been brought into service by the B.B.C. It replaces the temporary mobile station which has been in use at Glencairn for the past two years. The transmitter, which has an e.r.p. of 20 kW, shares Channel 1 (vision 45 Mc/s, sound 41.5 Mc/s) with Alexandra Palace.

It will continue to receive its programmes directly by radio from Kirk o' Shotts, Scotland. As has been done in the past, if the received picture is unsuitable for retransmission, or below standard, a warning signal—a vertical white bar—will be radiated. This warning signal is also being used by the Norwich transmitter which re-radiates London's transmissions.

The construction of the 640-ft self-supporting tower to carry the aerials for the new B.B.C. London station at Crystal Palace is well under way. At the 440-ft level will be installed parabolic aerials for receiving outside broadcasts. The Band I transmitting aerials—consisting of eight stacks of four dipoles—will be mounted on the section between 440 and 600-ft, above which there will be a 40-ft topmast. The tower is being erected by B. I. Callender's Construction and the aerial by Marconi's.

I.T.A. Links

A RADIO link is to be provided by the Post Office between Birmingham and the I.T.A. Midland station near Lichfield. It will provide two channels to Lichfield and one in the reverse direction. The Lancashire station at Winter Hill, near Bolton, will be linked with Birmingham by cable.

All the equipment for the radio link is being supplied by the G.E.C. who are also supplying the repeaters and terminal equipment for the cable link.

PERSONALITIES

Dr. Robert Cockburn, M.Sc., Ph.D., A.M.I.E.E., principal director of scientific research, guided weapons and electronics (M.o.S.) since March, 1954, has been appointed deputy controller of electronics in the Ministry in succession to Rear Admiral G. Burghard, whose tour of duty has expired. Dr. Cockburn, who is 44, was scientific adviser to the Air Ministry before joining the Ministry of Supply. For some time during the war he was head of the counter-measures group of T.R.E. and was awarded the American medal of merit for his work in this field.

Dr. A. L. Cullen, Ph.D., B.Sc., A.M.I.E.E., is to occupy the newly created chair of electrical engineering in the University of Sheffield. After graduating at the Imperial College of Science and Technology, London, in 1940, he joined the staff of the Royal Aircraft Establishment, Farnborough, where, until 1946, he worked on radar, being mainly concerned with aerials and waveguide techniques. He is at present reader in electrical engineering at University College, London, which he joined in 1946 as a lecturer. Dr. Cullen's special interest is microwave measurement techniques.

J. A. Smale, engineer-in-chief of Cable and Wireless since 1948, has been appointed by the government of Cyprus to be part-time chairman of the new Cyprus Inland Telecommunications Authority set up to administer and operate the island's inland telephone and telegraph services. He will continue in his present post with C. & W., visiting Cyprus as necessary.

H. R. Whitfield, M.I.E.E., who has been with Kelvin and Hughes, Ltd., as chief radar engineer since 1946, has been appointed a director of Kelvin Hughes (Marine), Ltd. In 1936, at the age of 19, he went to the Automatic Telephone and Electric Co., Liverpool, as a transmission laboratory engineer. From 1938 to 1940 he was a member of the War Department civilian technical staff attached to the Bawdsey (Suffolk) radar research station and then went to the research establishments, Malvern, as a member of the scientific staff on gunnery radar.

In addition to those mentioned in our last issue as having received Birthday Honours, John N. Toothill, general manager, Ferranti, Ltd., Edinburgh, was appointed C.B.E., Horace D. McD. Ellis, engineer in the B.B.C. Designs Department, was appointed M.B.E. and William Fairhurst, foreman, Electronic Tubes, Ltd., High Wycombe, received the British Empire Medal.

R. E. Burnett, M.A., A.M.I.E.E., A.Inst.P., has relinquished the principalship of Marconi College, Chelmsford, which he has held since 1950 and the position of manager of education and technical personnel on his appointment as full-time assistant to the general manager. His duties as manager of education and technical personnel will be undertaken by E. R. L. Lewis, M.A., A.M.I.E.E., who has been his deputy. The new principal of the college is R. G. Hulse, B.Sc., who has been deputy principal during the past year.

L. Hampson, who in this issue describes a Band II tuner unit, has been in the valve measurement and application laboratory of Mullard's since 1951. He joined Mullard's immediately after graduating at the age of 24 at Manchester University where he studied electrical engineering after completing his National Service. While at Mullards he has been mainly concerned with development work on the application of valves in the v.h.f. bands.

After 36 years' service with Marconi's, H. C. Van de Velde has relinquished the position of deputy to the managing director of the Marconi Marine Company but will continue to represent the company on the boards of various associated companies overseas. With Mr. Van de Velde's retirement, R. Ferguson, the general manager, who joined the seagoing staff of Marconi's in 1910, is extending his managerial responsibilities. He was seconded to the M.W.T. Company in 1934 in order that he might

take up the appointment of general manager of Egyptian State Broadcasting which Marconi's conducted for the Egyptian government. On his return from Egypt he was re-appointed to the marine company, of which he became general manager in 1947.

F. G. Robb, chief of Marconi's Test Division since 1948, has retired after 36 years' service with the company. He was for some years in the designs and development section where at one time he worked on the development of beam transmitters for the Marconi-Franklin short-wave beam system. During the war he was seconded to the Admiralty and became chief of radar





R. FERGUSON

E. H. EVANS

test. His successor at Chelmsford is **E. H. Evans** who joined Marconi's in 1913. He has been associated with the test division throughout his service and for a number of years has been chief of receiver test.

Obituary.—The death occurred on July 9th of L. F. Fogarty, M.I.E.E., who was for many years honorary treasurer of the Wireless Society of London, of which he was a founder-member and then of the Radio Society of Great Britain as it has been known since 1922. He had been managing director of the Zenith Electric Company since its formation in 1918.

IN BRIEF

Receiving Licences Decrease.—Although there was a further increase of 43,192 television licences in the United Kingdom during May, there was an overall decrease of some 17,000 in the number of domestic receiving licences in force. The comparative figures for May and April—the latter in brackets—are: sound 9,102,995 (9,165,242), vision 4,623,917 (4,580,725), car radio 273,883 (271,480), total 14,000,795 (14,017,447).

G. A. Briggs, of Wharfedale Wireless Works, Bradford, whose two lecture-demonstrations in the Royal Festival Hall, London, have created a demand for tickets far in excess of the seating capacity of the hall (3,000), is going to New York to give a similar demonstration in the Carnegie Hall (seating 2,760) on October 9th. Capitol Records Inc. are to make the recordings required for comparing live and recorded performances of piano, violin and organ and, as at the R.F.H., Mr. Briggs will be working in collaboration with P. J. Walker.

At the annual general meeting of the Television Society, the following members were elected to fill the vacancies on the Council: T. W. Price (Ediswan), A. E. Sarson (Marconi's), W. R. Smith (G.P.O.), Professor Trewman (E.M.I. Institutes) and C. B. Townsend (G.E.C.). In addition, D. N. Corfield (S.T.C.) and F. Livingston Hogg (Livingston Laboratories), who were co-opted last year, were elected full members.

B.A.T.C.—A convention of the British Amateur Television Club is being arranged for October 1st at the Bedford Corner Hotel, Bedford Square, London, W.C.1,

from 10 a.m. to 6 p.m. There will be a display and demonstration of members' equipment and a film show. Tickets (costing 5s) and further information can be obtained from D. S. Reid, 4, Bishop Road, Chelmsford, Essex.

Telesurance Limited, which operates a television insurance and maintenance scheme through registered R.T.R.A. dealers, has issued a statement on its policy regarding sets converted for Band III. No additional premium charges will be made providing the sets are converted by appointed dealers in accordance with the recommendations of the manufacturers.

The mains and output transformers and the smoothing choke specified by W. A. Ferguson for the 20-watt quality amplifier described in our May and June issues, are being produced by Partridge Transformers, Limited, Roebuck Road, Tolworth, Surrey. A leaflet giving electrical and physical characteristics is available.

In preparation for the advent of commercial television a television training centre has recently been opened in London by Marconi's. It provides a complete training course in the operation and maintenance of television studio equipment. The centre is also available to organizations for rehearsals under operational conditions and a mobile unit is available for the production of recorded O.B.s.

Engineering Education.—Lists of colleges in London and the Home Counties providing engineering courses during the 1955/6 session are given in "Engineering Education in the Region" published by the Regional Advisory Council for Higher Technological Education. It includes sections covering engineering crafts (including radio and television servicing), City and Guilds courses in telecommunications engineering, H.N.C. courses in electrical engineering with a bias towards radio and telecommunications, and courses in direct preparation for I.E.E. and Brit.I.R.E. examinations. The 30-page booklet is obtainable (price 1s) from the Regional Advisory Council, Tavistock House South, Tavistock Square, London, W.C.1.

At the eighth annual presentation of diplomas at the College of Aeronautics, Cranfield, Bucks, where a chair in aircraft electrical engineering has now been established, nine of the 68 awards were gained by students specializing in aircraft electrical engineering. The theses covered work on computors, simulators, servomechanisms and on the properties of suppressed aircraft aerials. As already announced, the first professor of the new department is G. A. Whitfield, who was at R.A.E. Farnborough.

"The Inquiring Mind" is the title of the documentary film sponsored by the I.E.E. for the purpose of depicting the diverse fields of opportunities open to electrical engineers. Copies of this 30-minute monochrome sound film, which is available in 35 mm and 16 mm, can be borrowed by schools, colleges and similar establishments. Particulars regarding the loan of the film are obtainable from the I.E.E., Savoy Place, London, W.C.2.

Nuclear Electronics.—During the forthcoming international conference at Geneva on the peaceful uses of atomic energy (August 8th to 20th) to which 84 nations have been invited, two exhibitions are being held. At one of these, which will be open to the public, a number of British electronics manufacturers will be participating and there will be a combined exhibit organized by the Scientific Instrument Manufacturers' Association.

R.E.C.M.F. Headquarters.—The Radio and Electronic Component Manufacturers' Federation has moved from Surrey Street, London, W.C.2, to 21, Tothill Street, London, S.W.1. (Tel.: Abbey 4226/8.)

Swindon.—With the object of forming a radio club in the town a meeting will be held on August 31st at 7.30, at the Connaught Café, 34, Cromwell Street, Swindon. Further particulars are obtainable from R. Reynolds, G31DW, 136, Beech Avenue, Swindon, Wilts.

Many demonstrations of high-quality reproduction have been given to gramophone and music societies and at public exhibitions by Goodmans Industries during the past few months. Societies interested in receiving a visit from the demonstration team are invited to write to Goodmans Industries Limited, Axiom Works, Wembley, Middx.

The third edition of the CABMA Register (1955/6) of British products and Canadian distributors again incorporates an alphabetical buyers' guide to some 4,000 British products, including radio and electronic equipment, available on the Canadian market. Other sections list manufacturers, trade names, Canadian distributors, etc. The register (price 2gns) is published jointly by Kelly's Directories and Ilifte and Sons for the Canadian Association of British Manufacturers and Agencies which operates British Trade Centres in Toronto, Vancouver and Montreal.

EXHIBITION NEWS

Next Physical Society Exhibition.—It is announced by the Physical Society that next year's 40th exhibition of scientific instruments and apparatus will be held in both the Old and New Halls of the Royal Horticultural Society, Westminster, London, S.W.1, from May 14th to 17th.

Marine Exhibition.—A number of radio manufacturers specializing in marine equipment or industrial electronic gear will be among the 500 exhibitors at the Engineering, Marine and Welding Exhibition which opens at Olympia, London, on September 1st for 13 days. Manufacturers appearing in the provisional list of exhibitors include B.T.H., Decca, English Electric (Electronics Division), G.E.C., I.M.R.C., Metropolitan-Vickers, Mullard, Radio Heaters, Redifon and Stratton. The exhibition will open daily from 10 a.m. to 8 p.m.

The dates for the ninth annual Amateur Radio Exhibition, organized by the Radio Society of Great Britain, have now been confirmed; November 23rd to 26th. It will again be held at the Royal Hotel, Woburn Place, London, W.C.1, and will be opened at noon on the 23rd by Vice-Admiral J. W. S. Dorling, director of the Radio Industry Council.

Midlands Radio Show.—The success of last year's Nottingham Radio Exhibition has promoted the organizers—the Nottingham Centre of the Radio and Television Retailers' Association—to broaden the scope of this year's show and to re-name it the Midlands Radio Exhibition. It will be held in the Ice Stadium, Nottingham, from September 19th-24th.

At the British Exhibition in Copenhagen (September 29th to October 16th) arranged jointly by the British Import Union, Denmark, and the Federation of British Industries, a large stand has been taken by the Radio and Electronic Component Manufacturers' Federation. Some twenty member-firms will be participating.

The Model Engineer Exhibition, with which is combined this year the Exhibition of Inventions, is to be held at the new Horticultural Hall, Westminster, from August 17th to 27th.

BUSINESS NOTES

Collins Radio, the well-known American manufacturers of aeronautical radio equipment, have formed a subsidiary in this country—Collins Radio Company of England, Limited. At present it is operating a service depot at Sunflex Works, Colham Mill Road, West Drayton, Middx. (Tel.: West Drayton 2226.)

Ekco-Dynatron Merger.—E. K. Cole, Ltd., have acquired a controlling interest in Dynatron Radio, Limited, of Maidenhead, Berks. It is understood that there is no intention of changing the policy of the business which is concerned with the manufacture of "above-average" domestic sound and television receivers.

The potential output of G.E.C. television receivers will be increased by 50 per cent with the rearrangement of production facilities in Coventry. The factory in Spon Street is being devoted exclusively to television and the production of domestic sound receivers is being transferred to another factory in the city.

The exclusive world distribution of acoustical equipment developed by **Kelly Acoustics Limited**, of 295, Regents Park Road, London, N.3, has been taken over by **Thermionic Products Limited**, Hythe, Southampton, to whom all enquiries for the new ribbon loudspeaker (RLS/1) should be sent.

Ampex Corporation, of California, manufacturers of magnetic-tape recording equipment, are to open an office in London. This and similar offices in overseas countries will be run by the recently formed company, Ampex International.

The "Radiovoyce" microphone equipment illustrated on p. 312 of our July issue, for which Leevers-Rich Equipment Limited are sole marketing agents to the trade, is designed and manufactured by F. W. Hopwood (Developments) Limited, 181, Wollaton Street, Nottingham.

Truvox Limited announce that they have appointed A. B. Thompson (Ireland) Limited, of 15, Newforge Lane, Belfast, as Northern Ireland agents for their tape-recording components and accessories.

F. W. Electronics, Limited, of New Southgate, London, N.11, which was formed in 1950 and specializes in the design and manufacture of audio and r.f. equipment—including equipment for schools—has moved its works and registered office to 12a, Prince of Wales Road, Hendon, London, N.W.4. (Tel.: Sunnyhill 0683.)

Holiday and Hemmerdinger, Limited, of 74-78, Hardman Street, Deansgate, Manchester, 3 (Tel.: Deansgate 4121) have notified us that they have arranged with the Recording Equipment Division of E.M.I. Sales and Service, Limited, to distribute "Emidisc" lacquer recording blanks and accessories to the trade in the Manchester area.

The telephone number of Nagard, Limited, designers and manufacturers of electrical instruments for research and industry, of 18, Avenue Road, Belmont, Surrey, has been changed to Vigilant 9161.

The North-Western Gas Board, which operates two base stations at Manchester and Liverpool and 25 mobile radio-telephone installations, is now using a radio-equipped

radio-telephone installations, is now using a radio-equipped mobile "paying-in" office.

The a.m. equipment being used in the vehicle, which periodically visits outlying districts, was supplied by the Radio & Transmission Division of Automatic Telephone & Electric Company. Two additional fixed stations are to be brought into operation by the Board to serve the Wirral and St. Helens areas.

Jack Davis (Relays), Ltd., is moving this month from Percy Street, London, W.l, to Tudor Place, Tottenham Court Road, W.I. Telephone numbers are unchanged.

HALF A MILLION television receivers have come off this assembly line at the Enfield. Middlesex, factory of Ferguson. The conveyors in the foreground are carrying completed chassis, after soak tests, to the section for final testing and adjustment.

EXPORT NEWS

Equipment for a public radio-telephone service in the three neighbouring territories of Sarawak, North Borneo and Brunci is being provided by A.T.E. (Bridgnorth), Ltd., a subsidiary of Automatic Telephone & Electric Company, Ltd. More than 80 single-channel radio links will be required for the service which will link the outlying settlements and the divisional centres where line telephone services already exist. Eventually the main centres will be linked by a multi-channel radio system.

Redifon radio-telephone equipment is being fitted in ten of the vessels of the Niger river fleet of the United Africa Company and fixed stations will be set up at **B**urutu, Makurdi, Yola and Garua.

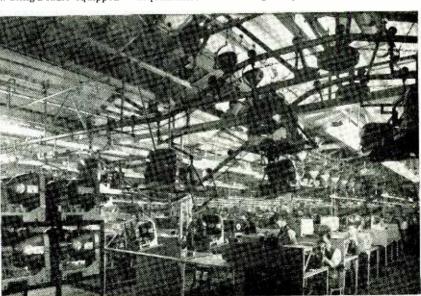
A radio network providing for 240 telephone circuits and a two-way 525-line television channel between Osaka and Fukuoka has been ordered from Standard Telephones and Cables through its associates the Nippon Electric Company, Limited, of Tokio. Eleven intermediate repeaters, working in the s.h.f. band (3,000-30,000 Mc/s), will be used to cover the 385 miles.

India's Director General of Supplies and Disposals (Shahjahan Road, New Delhi), has asked for tenders for 12,500 broadcast receivers and associated aerial equipment and loudspeakers. The majority of the receivers are for dry-battery operation and must cover the medium-wave band although some of them must also cover the short waves. About 1,000 receivers are needed for a.c. operation. A copy of the tender documents is available from the Export Services Branch, B.o.T., Lacon House, Theobalds Road, London, W.C.1 (Reference ESB14288/55). Closing date for tenders is July 29th.

International Aeradio, Limited, has received from the Egyptian Air Force an order for three air traffic control desks. I.A.L. is also providing the radio and air traffic control services at Yeadon Airport, near Leeds, which is to be developed to provide scheduled and charter air services.

Airfield control radar equipment (Type 424) is to be supplied by **Decca Radar**, **Limited**, for installation at Durban National Airport. This surveillance radar equipment, which was reviewed in our November, 1953, issue, was also recently supplied to the South African Air Force.

R. B. Page, of Birmingham Sound Reproducers, Ltd., is on a three-month visit to North America to renew acquaintance with radio-gramophone manufacturers.



359

Inexpensive Wave Analyser

"Zero-beat" System Using Simple Low-Pass Filter

By M. G. SCROGGIE, B.Sc., M.I.E.E.

N a recent survey of distortion-measuring technique¹ it was noted that published data on distortion usually take the form of a single figure ("total harmonic distortion") whereas for fair comparison one must know about the individual distortion products (whether harmonics or intermodulation) making up this total. One reason for the scarcity of analysed data is no doubt the high cost of wave analysers. In order to measure each distortion product separately it is necessary to have extremely high selectivity, which cannot reasonably be obtained by straightforward a.f. tuning capable of being varied continuously from, say, 20 c/s to 20 kc/s. The difficulty is usually overcome in the same way as in the analogous r.f. problem in radio receivers, with the aid of the superheterodyne principle. By means of a beat oscillator, the frequency of the chosen component of the signal being analysed is transferred to an "i.f.," which might be 50 kc/s; components of all other frequencies are then removed by a filter having a pass band of only a few cycles per second, and after amplification the selected signal deflects an indicating

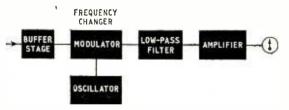


Fig. 1. Block diagram of the type of wave analyser described.

The heart of the instrument is the filter, because on it the capabilities of the wave analyser chiefly depend. For testing high-fidelity equipment it is necessary to measure distortion components of the order of 0.1% (i.e., 60dB down) relative to a fundamental output separated in frequency by perhaps 25 c/s, which relative to 50 kc/s is 0.05% off-tune. At the same time, in order not to render the analyser too tricky to use, or make unreasonable demands on signal-frequency stability, the filter characteristic ought to be flat-topped. Such onerous requirements, calling for carefully applied crystal resonator technique, have no doubt deterred many experimenters from running up an analyser for themselves.

Continuing the radio receiver analogy, we might remember that there is such a thing as a synchrodyne, ^{2,3} which can be defined as a superhet in which the i.f. is zero. An advantage therein is that instead of the usual highly selective band-pass i.f. system a simple low-pass filter will do. This seems just what is wanted for a wave analyser. Providing a filter to cut out everything above one or two cycles per second presents no difficulty at all. In fact, on top of a

great saving in expense, it is easy to obtain substantially higher effective selectivity than in a conventional wave analyser, yet at the same time adjustment is less critical.

One inherent disadvantage of the principle ought perhaps to be declared at the outset. In a frequency changer, harmonics of the oscillator frequency are likely to be present. These, in a wave analyser with an i.f. of the order of 50 kc/s, are too far up in the r.f. region to give i.f. beats with any a.f. (the "bias" frequency in a tape recorder must be watched, however). But in the zero-i.f. type of frequency changer, in which the desired response is obtained by setting the beat-oscillator frequency very nearly equal to that of the signal component to be measured, smaller responses can also occur at multiples (especially odd multiples) of the frequency read. One must therefore take some care to choose signal frequencies that cannot yield misleading responses.

What might be considered to be another disadvantage is that if one desires to cover the full a.f. band the beat oscillator has to be variable over a large tuning ratio, say 1000:1, as compared with less than 2:1 in the conventional system. On the other hand, however, the fact that the beat oscillator frequency at zero-beat is exactly the same as that of the component being measured makes for much easier adjustment and more accurate and stable frequency calibration, which are decided advantages.

Fig. 1 is a block diagram of a zero-i.f. wave analyser. When measuring very weak distortion products, some amplification ahead of the frequency changer could be helpful, but it is even more important not to add to the distortion before it is measured. Some sort of buffer stage is essential, in order to make the input impedance large enough not to impose appreciable loading on the signal being analysed and at the same time to present to the modulator an output impedance low enough not to introduce appreciable distortion. These requirements are met by a low-resistance cathode follower.

Type of Oscillator

The modulator is a bridge comprising four rectifiers, shunted diagonally across the signal path. The design of the beat oscillator, which is connected across the other diagonal, depends on requirements. For experimental purposes an a.f. generator previously described was more than adequate, and enabled results using square waves to be compared with those using sine waves. For routine tests along the lines suggested, however, a simple fixed-frequency oscillator could be used. A 3-valve oscillator like either of the two shown in Fig. 6 is very suitable for either fixed or variable frequency.

Because the mechanical movement of the indicator

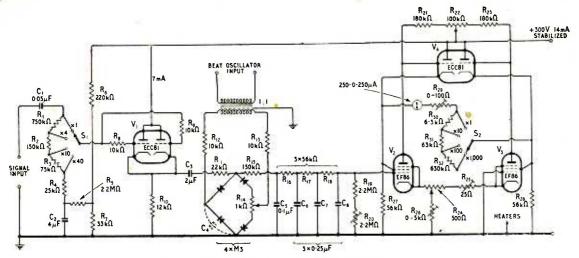


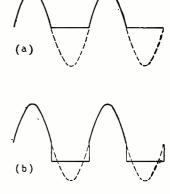
Fig. 2. Circuit diagram of a particular development of Fig. 1.

is itself a low-pass system, a very simple and cheap resistance-capacitance filter is all that is needed between modulator and amplifier. It was this amplifier that presented the greatest difficulty. At first is was supposed that two resistance-coupled stages with the time constants adapted to pass a band of the order of 0.2 to 2 c/s, or a balanced amplifier with heavy negative feedback except over this band, would be a suitable basis for design. By cutting out zero frequency it was hoped to obtain a stable meter zero, but it was found that even in a push-pull system it was impracticable to balance the long time constants sufficiently to avoid perpetual slow drifts, and in spite of the zero-instability problem a d.c. amplifier was actually more workable. Stabilized power supplies are of course necessary in any case.

The design can now be discussed in detail with reference to the circuit diagram, Fig. 2, which is subject to modification to suit individual needs. In order to provide a low output resistance (about $130\,\Omega$) at a moderate anode current (7mA) the cathode follower comprises both halves of a high-slope double triode. The distortion caused by it is too small to estimate precisely; but the total residual intermodulation, which with two input signals of different frequency, each 2½V peak, is of the order of 0.05% is perceptibly increased if only one half is used. Without R_8 and R_9 there is a tendency to parasitic oscillation. The time constant C_2 R_5 has a valuable stabilizing effect in preventing short-term fluctuations in h.t. voltage from reaching the grid. C2 has in any case to be fairly large for its impedance at the lowest input frequency (reckoned as 20 c/s) to be negligible in relation to R4, the lowest step of the input potential divider, which functions as a scale multiplier. Obviously the switch controlling it, S₁₃ should be of the make-before-break type.

The modulator (known as a Cowan type ^{2,3}) is made up of four Standard Telephones M3 miniature selenium rectifiers. Under the influence of the beat oscillator their resistance becomes alternately much higher and much lower than R₁₁, so that the signal from the cathode follower is alternately passed to the filter and suppressed. When the signal contains a frequency exactly equal to that of the oscillator, and in phase, the component of the signal at that frequency

Fig. 3. Signal waveforms across the modulator, (a) according to simplified theory, and (b) as modified in practice by C_3 (Fig. 2). The dotted portions of the original waveform are suppressed. In (b) the horizontal portions indicate the displaced zero level.



is in effect rectified. If the phase is reversed, the direction of rectification is reversed. So if the signal frequency is altered by, say, 1 c/s, a 1-c/s signal is created by the slowly shifting phase relationship. This signal is accepted by the filter, the original signal being rejected.

Modulation Wave Shape

If the action were quite as simple as just described, the waveform across the modulator, with the beat oscillator in phase with a sinusoidal incoming signal, would be as in Fig. 3(a), where the dotted line traces the suppressed half-cycles. This condition corresponds to the peak value of the beat-frequency signal, which is clearly equal to the mean value over a whole cycle of one half cycle of the incoming signal, or $1/\pi$ times its peak value. The theoretical efficiency, neglecting all losses, such as those due to the finite resistance of the rectifiers, would therefore be barely 32%. In reality, however, C_3 modifies the action in a manner analogous to that of the reservoir capacitor in an ordinary rectifier circuit, displacing the waveform upwards and increasing its mean value and To obtain the greatest therefore the efficiency. benefit from this, the beat oscillator should have a pulse waveform, so that (just as in the rectifier circuit) the charging of C₃ is concentrated at the peak

of the input signal. After investigation it was decided that the increase in efficiency was not worth the trouble of providing the special beat-oscillator waveform and ensuring constancy of its positive/negative time ratio at all frequencies. The undesired responses are also more liable to be troublesome, because if the waveform ratio is m:1 the only harmonics suppressed are integral multiplies of m+1. Using a 1:1 waveform thus suppresses all even harmonics. Both for this reason, and in order to obtain constant efficiency, it is advantageous to preserve an accurate 1:1 ratio. This is easier to do with a sinusoidal waveform than with a square; so although the square gives the quickest switch-over for a given peak value, and consequently slightly lower residual inter-modulation than an equal sinusoidal voltage, the latter was chosen on balance. The resulting signal waveform is as Fig. 3(b), and the measured efficiency (inclusive of the cathode follower) 52%. For comparison, the efficiency with a 4:1 oscillator waveform was 75%.

The time constant C_3R_{11} is important, because it must be long compared with the period of the lowest signal frequency and short compared with that of the highest beat frequency. Since the optimum R_{11} depends on the backward and forward resistances of the rectifiers, a suitable time constant is obtained

by choice of C₃.

The higher the oscillator voltage the greater the signal voltage that can be handled linearly, but of course it must be within the maximum rating for the rectifiers, which is 56V peak inverse per rectifier, or 112 for the bridge, less an allowance for inequality of backward resistance. The signal voltage being read is limited by the amplifier to about 10V peak, and for that a sinusoidal oscillation of 20V r.m.s. is sufficient. To allow a margin of amplifier linearity, the maximum reading has been reckoned as that which is given by $6V\ r.m.s.$ at the input with S_1 at $\times 1$. Provided that S_1 is not moved to a more sensitive setting than that at which the strongest signal component is within the 6-V limit, the peak value of the whole signal at the modulator is not likely to exceed about 20, and using 28V r.m.s. (40V peak) for modulation it has been checked that this amount of signal voltage does not appreciably affect the accuracy of reading (e.g.) a 4mV component of it. It may, however, make it rather less easy to read, and 10V peak is a more conservative limit. The mean rectified current in the transformer secondary at 28V r.m.s. is about 0.5mA, which is well below the rated maximum of 1mA per rectifier. Linearity of the modulator is excellent.

Meter Zero Stabilization

In theory, this type of modulator automatically balances out the rectified and oscillator-frequency voltages from the signal path, but owing to inevitable inequalities in rectifier characteristics there is in practice a residue of both. A reasonable amount of this has not been found to cause appreciable error, but even a very small residual fraction of the total rectified (z.f.) voltage is enough to displace the pointer considerably. That in itself could be taken up on the amplifier balancing adjustment (R₂₄), but unfortunately the z.f. residue tends to vary with oscillator frequency, so that every change of frequency necessitates readjustment of meter zero. This nuisance can be more or less eliminated by (1) winding the

transformer in the manner recommended for impedance bridges-with the secondary in two identical halves, balanced with regard to earth (represented mainly by the inter-winding screen);* (2) using the balance control R_{14} ; and (3) connecting a capacitance C4 across one of the rectifiers. This capacitance, of the order of 100pF, has most effect at high frequencies. The procedure is to adjust R₁₄ so that switching on the beat oscillator at some medium or commonlyused frequency (say about 400 c/s) causes no displacement of the meter on the most sensitive range; then to vary the oscillator frequency and try various values and positions of C4 to minimize shift when the frequency is raised. One of the advantages of tests at a single frequency1 is that no special transformer or other precautions are needed. Also the advantages of a square modulating waveform can be obtained without most of the disadvantages.

The component values of the filter are not critical, but obviously R_{15} must be large compared with R_{11} if efficiency is not to be lost. As an example of the selectivity obtained by the simple means of this instrument, reading an input of 2.5mV at any frequency is quite unaffected (except for a very slight vibration of the pointer) by the presence of 4,200mV

(i.e., 65dB stronger) only 30 c/s away.

Negative Feedback

At first glance the amplifier circuit may look rather like that of the valve voltmeter described in the August 1954 issue, but whereas in that circuit the whole voltage gain was sacrificed in the interests of stability by connecting the output terminals straight back to the control grids of V2 and V3, here a large voltage gain is needed in order to be able to measure small distortions, and the only negative feedback is that which results from feeding the screen grids of V₂ and V₃ from the output terminals. This policy not only saves a special potential divider for the purpose, but it provides some degree of zero stabilization, and lowers the output resistance of the amplifier, giving a more stable calibration. The resistance of the $250-0-250\,\mu\text{A}$ meter is $500\,\Omega$, and the output resistance of the amplifier about $160\,\Omega$; R_{29} is used to bring the whole up to a level of $700\,\Omega$, so that R_{30} R_{32} with the values stated give decade ranges by means of S_2 .

 R_{26} is used to set the valves to suitable working points, indicated by the total amplifier h.t. current being about 7mA. R_{24} is used as a coarse balancing or zero-setting control and R_{25} the fine control. The additional balancing facility afforded by R_{22} is not absolutely essential, but is quite helpful in arranging

the best working condition.

Without R_{19} and R_{20} the system was found to give full-scale swing on the $\times 1$ ranges of S_1 and S_2 for 9mV r.m.s. input at the selected frequency. By means of R_{20} this is pre-set to a convenient 10mV, the scale having been provided with a 10-0-10 marking. On the $\times 1,000$ setting of S_2 the meter is therefore direct reading in volts, but (as already explained) should not be used above 6 on this range.

The overall voltage gain of the amplifier is thus about $\times 24$. Although not large by higher-frequency standards it can cause quite a lot of trouble unless care is taken with regard to zero stability. The balanced circuit of course goes a long way towards

^{*} A suitable transformer can be obtained from the Majestic Winding Co., 180 Windham Road, Bournemouth.

achieving this and those inexperienced with d.c. amplifiers may perhaps wonder why it does not go all the way, since it might appear that any change in supply voltages would affect both halves of the system equally and therefore would not affect the meter. But a little calculation shows that a perceptible displacement of the pointer—say $10\mu A$ —results from a difference between the anode currents of V_2 and V_3 of only about $0.05\mu A$. Now although any initial difference can of course be corrected by R_{24} or R_{22} it would be too much to expect the anode currents of even a well-matched pair of valves to vary equally within $0.05\mu A$ over a range of anode, screen and heater voltages.

Any reasonably effective stabilizer for the source of the 300V h.t. should be able to eliminate zero-shift due to variations in anode and screen potentials of valves V₁-V₄, but the provision of R₂₂ in addition to R24 makes assurance doubly sure; by successively shifting the setting of one and recentring the meter pointer with the other, an adjustment can be found at which even several volts change in h.t. has little effect on the meter, and protection is thus obtained against transient fluctuations of voltage of an inferior stabilizer. Protection against such transients (but of course not against long-term drifts) via the grids of V, is given by C₂R₅. Trouble due to small fluctuations of h.t. is thus confined almost exclusively to the cathode current of V₁, and it is for this that stabilization of the supply is chiefly needed. Even here an exceptionally high standard is not essential, thanks to the stabilizing effect of the negative feedback.

Incidentally, in case it occurs to anyone to extend the balanced-circuit principle all the way, as in Fig. 4, it should be mentioned that this causes the full signal voltage from V_1 to be applied between the grids of V_2 and V_3 and earth, thereby prematurely overloading

V₂ V₃ V₃ V₄ V₄ V₄ V₅ V₇ (6)

Fig. 4. Alternative, but less satisfactory, arrangement of the circuit up to the d.c. amplifier.

the amplifier on its most sensitive range. Moreover differences in time constants cause h.t. fluctuations to be more, rather than less, troublesome.

Although changes in oscillator amplitude affect the meter by upsetting the adjustment of R₁₄, it has not been found necessary to stabilize the oscillator power supplies, sufficient control being provided by the thermistor.⁴

Except for very low frequency fluctuations that may come in with the signal, the remaining cause of zero instability is the heater voltage. Its effect on V_4 is slight. On V_1 it is more troublesome; a 5% change in V_h was found to cause 100μ A zero shift on the most sensitive range. But the effect is most serious as regards V_2 and V_3 . The simple device of differentially adjusting the heater voltages of the two valves, for which very good results have been claimed with triodes, was found to be quite ineffective, at least as regards the few samples of EF86 available. If one is not so fortunate as the writer to pick two valves that are well matched, not only as regards the ultimate

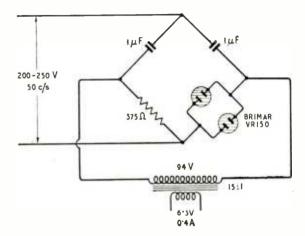


Fig. 5. Cherry and Wild stabilizer for the heater supply to V_2 and V_3 , if found necessary. The VRI50 tubes in parallel are connected anode to cathode.

inequality in anode-current change caused by a heater-voltage change, but also in the rate at which the anode-current change occurs, it is advisable to stabilize the heater voltage of at least these two valves. If it can be stabilized for all the heaters, so much the better; but stabilization of V₂ and V₃ only is suggested as a second-best, because the comparatively simple Cherry and Wild system⁶ provides sufficient output for these two heaters. Fig. 5 shows the circuit, which was found to be amply effective for badly matched valves, giving a stabilization ratio of the order of 30:1.

Type of Meter.

Heater-voltage stabilization is rendered ineffective if there is instability of heater-circuit resistance, so good valve-holder contacts are essential; it would be better, if suitable valves with wire leads were obtainable, to solder them in without holders. Of course the whole of the circuit, especially from signal input to the grid of V₂, must be free from uncertain contacts, leakages and stray pick-up; for example, all capacitors should have high and constant insulation resistance, and their cases should be earthed.

Lastly, there is the indicator itself. A 2-in dia.

0-500 µA meter, adjusted to a centre zero, was found to have a suitable combination of electrical and mechanical characteristics. Owing to the automatic limitation of current by the circuit, a more sensitive instrument could be used without risk of accidental damage, but there would be no point in doing so, because the minimum effective reading is already limited more by instability of zero and residual distortion than by insensitivity of the meter. Damping of the movement must be enough to suppress resonance without causing excessive sluggishness. The higher the frequency at which the pointer can oscillate without serious error (10% or even 20% is a not unreasonable tolerance in distortion measurements), up to the rate at which reading its amplitude of swing becomes difficult for the eye, the less critical is the adjustment of the signal frequencies and the demands on oscillator frequency stability. In practice a beat period of several seconds per cycle is about right.

Intermodulation Twin Oscillator

Coming now to the signal source, the chief requirement is stability and ease of adjustment of frequency. Dependable frequency calibration is a great help, particularly in identifying and avoiding spurious For measuring harmonics, very pure waveform is obviously needed; this is not so necessary with intermodulation, the errors caused by harmonic content then being only of a second order of magnitude. The three-valve circuit forming the nucleus of the a.f. source already mentioned has proved to be simple and reliable, with quite phenomenal frequency and amplitude stability. A double oscillator for intermodulation measurements, hastily put together without any special care, using ordinary components, has often been set to give a slow beat, at, say, 0.3 c/s, and has continued to maintain this rate, without perceptible change, for hours on end. As regards purity of waveform, the second harmonic is about 0.25%, third 0.3%, and all others negligible. This performance was obtained notwithstanding that six old EF50 valves were substituted for the SP61 used in the original signal generator, whose harmonics are about half as great. The suitability of EF80 valves was checked by plugging them in, via adapters, in place of the EF50; no readjustment was necessary; the total h.t. consumption for the two oscillators was increased from 40 mA to 46 mA, and the harmonic content nearly halved. In every way, therefore, this circuit is eminently suitable as a signal source for intermodulation measurements in conjunction with the wave analyser. Fig. 6 is the circuit diagram.

For economy as regards the ganged inverse semilog rheostats, the frequency ratio on each range was

TABLE I.

	Oscilla	itor 1	Oscillator 2			
Range	Fre- quency	Capaci- tance	Fre- quency	Capaci- tance		
1	40-120 c/s	0.1μF	350-	0.0115μF		
2	350- 1,100 c/s	$0.0115 \mu \mathrm{F}$	1,100 c/s 1,000- 3,200 c/s	0.004µF		
3 4	3-10 kc/s 9-25 kc/s	1,330pF 450pF	3-10 kc/s 9-25 kc/s	1,330pF 450pF		

reduced from the 10:1 in the original model to about $3\frac{1}{2}:1$; this also slightly facilitated the all-important precision in frequency setting, but the fine controls (R_2) are chiefly relied upon for close adjustment. For the purposes in view, complete coverage of the full a.f. band by both oscillators was unnecessary, and the ranges are as given in Table I.

If lower frequencies were included it would be necessary to increase the values of the coupling

capacitors accordingly.

Apart from these modifications of the original design, concerned with frequency coverage, the only new consideration was the necessity for combining the two signals in the output, without giving rise to intermodulation. With the direct mixing system shown in Fig. 6, there is just perceptible intermodulation, and if one wants to avoid this one can substitute a bridge or hybrid-coil system, but the improvement was found to be hardly worth while. It can be shown that, given equal outputs from the two oscillators, the ratio of their voltages at the common output terminals is equal to the ratio in which R₁₉-R₂₁ is divided by S₂ (where these resistances include the internal resistances of the oscillators, which, being cathode followers in this case, are negligible). therefore tests are usually made with either a 1:1 or 4:1 ratio, it is convenient to provide tappings as shown. This does not prevent any other desired ratio from being set up by the separate controls (R_{14}) ; the level of the combined signal can then be varied without change of ratio by R₂₂.

Avoiding Spurious Effects

The frequency-selecting section of each oscillator should be enclosed in a screen; it is particularly necessary to remove stray capacitance between the two sections, as failure to do so may cause a small signal from one oscillator to appear at the output even when its own R_{14} is at zero. No difficulty was found, however, in running the twin oscillator unit from the same stabilized power supply as the wave analyser.

Residual intermodulation of modulator, input cathode follower, and signal-oscillator output circuit was checked in turn by first feeding the modulator with a single 420c/s signal from oscillator 2 via analyser cathode follower in series with the secondary of a step-down mains transformer giving a 50c/s signal, each signal being adjusted to give a reading of say 2V on the analyser meter. Intermodulation signals were looked for at 420 ± 50 and 420 ± 100 c/s, and were barely perceptible on the most sensitive range. The 50c/s source was then transferred to the lead between signal oscillator and analyser cathode follower. Lastly, the test was repeated with the 50c/s mains source replaced by oscillator 1. Overall intermodulation using a sinusoidal beat-oscillator waveform, was 0.1% second order and 0.08% third order. With a 1:1square waveform the figures were 0.04% and 0.02% respectively.

Of spurious responses, the most important are those resulting from signals at multiples of the frequency being read, which is the frequency of the beat oscillator, denoted by f_b . Using a 1:1 waveform ratio, the responses due to signals at even multiples of f_b are theoretically absent, and those at odd multiples (frequency = nf_b) are one nth of the amplitude resulting from an equal amplitude signal at f_b . In practice this is approximately true of the odd multiples, but although the even-multiple responses are relatively

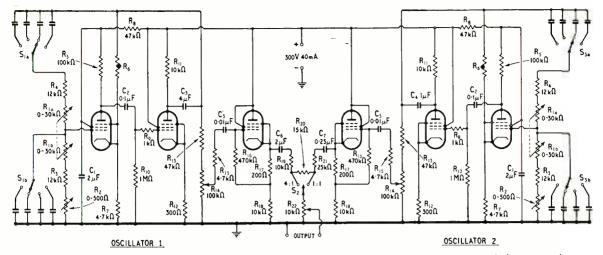


Fig. 6. Circuit diagram of a twin signal source for making intermodulation measurements in conjunction with the wave analyser. The thermistors (R_6) are Standard Telephones type A5513/100, and the valves all SP61, EF50, EF80 or equivalent. S_1 and S_3 are 2-pole 4-way frequency-range switches; the capacitances selected by pole b are the same as by pole a, and typical values are given in Table 1. Components in Oscillator 2 having the same values as the corresponding components in Oscillator I bear the same numbers.

small (<5% of the true response) their existence ought not to be overlooked.

When analysing signals it is inevitable sometimes that the component being read will be accompanied by another component at a multiple of its frequency, in which case an error must result. Fortunately it is seldom large enough to be appreciable. One reason for this is that, taking the odd and even harmonic series separately, their amplitudes almost invariably decrease more or less rapidly with frequency; another is the decreasing response of the analyser, which almost wipes out the even series and greatly discriminates against the odd.

In the very unlikely event of a sixth harmonic not being small compared with a second harmonic, it might (being three times the frequency) appreciably affect the reading of the second. But the only troublesome situation that is at all likely in practice is the measuring of a fundamental accompanied by a very large percentage of third harmonic. This situation is revealed by the non-sinusoidal swinging of the pointer, for the third harmonic in the signal is represented (though at only one third the amplitude) by a third harmonic of the slow beat frequency, and the effect of this harmonic can be still further reduced by bringing the fundamental beat frequency to the point beyond which response falls off rapidly.

In choosing f_1 and f_2 , the lower and higher signal oscillator frequencies for intermodulation tests, one would of course avoid making any of the components to be read— $f_2 \pm f_1$, $f_2 \pm 2f_1$, etc.—equal to a multiple of f_1 or f_2 . Fractional relationships between f_1 and f_2 would also be avoided; e.g., if f_1/f_2 were 2/5, $f_2 + 2f$ would clash with $f_2 - f_1$, being its three-fold multiple. Doubt about whether or not a response is due to intermodulation can be dispelled by checking that the beat frequency responds to an adjustment of both f_1 and f_2 .

Pick-up from the mains is another possible cause of undesired responses, and 50c/s and its multiples (especially odd) are to be avoided as signal frequencies.

It has been interesting, in using this wave analyser, to confirm the experimental measurements by Warren and Hewlett7 of the ratio of intermodulation to harmonics with amplifiers having various kinds of distortion and level frequency characteristics. Since the wave analyser reads only the sum or difference frequency and not both at once, the ratios are half those obtained by Warren and Hewlett with the same relative signal amplitudes (which were in the ratio 5 for harmonic to 4 and 1 for intermodulation measurement). With a single triode, giving almost pure second-order distortion, the amplitude of the component at $f_2 + f_1$ or $f_2 - f_1$ is 1.6 times that at $2f_1$; and with a push-pull amplifier, giving mainly third-order distortion, the amplitude ratio of $f_2 + 2f_1$ or $f_2 - 2f_1$ to $3f_1$ is 1.92. When negative feedback is used, distortion is of course greatly reduced; but if the signal amplitude is then raised to reintroduce distortion, its onset is more rapid, and instead of being concentrated mainly in second or third or both, it continues far up the series with comparatively slow convergence, and the aural unpleasantness is worse.

While admittedly a keen ear is the only instrument that is valid in the final assessment of sound reproduction, instrument readings are of very great value if properly taken and interpreted; and if analysed measurements rather than lumped were more generally made the much-needed establishment of accepted correlation between instrument readings and listening tests would assuredly be speeded up.

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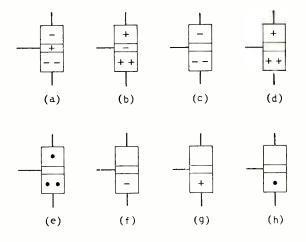
TRANSISTOR SYMBOLS

SUGGESTIONS for the symbol to be used for the transistor have appeared from time to time in Wireless World, and especially in the April and May issues (pp. 151 and 201). Another symbol of an interesting kind is used in L'Onde Electrique (March-April, 1955, pp. 243-263). The symbols for n-p-n and p-n-p junction transistors are shown in the figure, at (a) and (b) respectively. Plus signs are used to designate p-type material and minus signs for n-type.

Since the emitter, as their source, contains more charge carriers than the collector (by those leaving through the base), the emitter and collector are distinguished by two signs and one sign respectively.

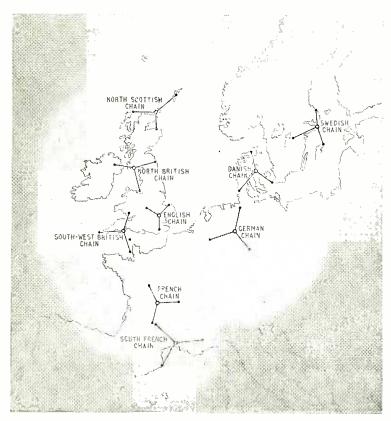
The full sign complement of (a) and (b) is actually unnecessary, for the base material is always of opposite kind to the emitter and collector. One can, therefore, omit the signs on the base and reduce the symbols to the form (c) and (d). This is quite commonly done in the article in L'Onde Electrique. A further modification consists in replacing the signs by dots as in (e). This is done when one wants to designate a transistor without being specific about whether it is a p-n-p or an n-p-n type.

There is still some redundancy, however, and the symbols could be reduced to (f), (g) and (h) without



losing anything. The convention here is to mark only the emitter by a sign indicating the nature of the charge carriers, so (f) is for an n-p-n transistor, (g) is for a p-n-p and (h) is a general symbol for either.

DECCA NAVIGATOR EXTENSIONS



PLANS have been made to build three new chains of Decca Navigator stations in Europe—in Sweden, southern France and northern Scotland—making a total of nine. When the new stations are completed some 2,000,000 square miles of Europe will be covered. As will be seen on this map, coverage will extend from Cape Finisterre to the Gulf of Bothnia and from Corsica to beyond the Faroes.

It is also announced that plans are being made for the first two chains of permanent stations to be erected outside Europe. They will cover the Bombay and Calcutta areas of India. Temporary lowpower chains have been set up overseas for survey purposes; one of the latest being for the Japanese Hydrographic Department.

Since the first Decca chain in south-east England was opened in July, 1946, over 2,600 naval and merchant ships have been fitted with the Decca Navigator. An increasing number of civil aircraft are also using it, and a new receiver, the Mark 10, designed especially for aircraft, is being produced. It will cover seventeen different chain frequencies and provide automatic lane and zone identification.

Wireless World, August 1955

LETTERS TO THE EDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

Spurious Radiations from Wrotham

IN his article in your July issue J. R. Brinkley drew general conclusions on the co-siting of f.m. transmitters from observations made on the performance of the transmitters at Wrotham as they are at present installed. The arguments against co-siting of television transmitters are of course irrelevant when applied to f.m. transmitters, and "co-siting" is perhaps a misleading term to apply to the practice of feeding one aerial system with the output from three transmitters.

Mr. Brinkley is correct in pointing out that adequate steps must be taken to reduce coupling between transmitters, and he is also right in his statement that at the present time the attenuation of the filters now installed at Wrotham is insufficient. But, because of the severe interference on the medium-wave band the B.B.C. decided, in agreement with the Post Office and the radio industry, to bring Wrotham into regular service before the second half of the Home Service transmitter and the aerial combining units for the three transmitters were in their final form. In this way a v.h.f. service was made available at the earliest possible moment to listeners in London and South-East England. The temporary filters at present in use will be replaced by the final filters within the next few weeks. When this has been done the intermodulation products to which Mr. Brinkley refers, and of which we can particular aware will become predictible.

we are naturally aware, will become negligible.

E. L. E. PAWLEY

Head of Engineering Services Group, B.B.C.

F.M. Receiver Design

WE cannot agree with J. K. Carter (correspondence in July issue) who rebukes us for using the ratio detector in our f.m. tuner described in the April and May, 1955, issues. The decision to use this in preference to the Foster-Seeley discriminator was made after considerable thought and the saving of one valve is only one of the factors which influenced us. A more important consideration is that the low distortion of the Foster-Seeley circuit can be obtained only by critical adjustment of the coupling between primary and secondary windings of the discriminator transformer. To make this adjustment requires equipment unlikely to be possessed by the amateur constructor. The linearity of the ratio detector is less dependent on circuit adjustment; moreover there is less inter-station noise than with the Foster-Seeley type.

Mr. Carter accuses us of being illogical but he con-Mr. Carter accuses us of being higher but he conveniently overlooks the other sources of distortion in the complete f.m. chain and in particular the chief offender; namely, the moving-coil loudspeaker. "Or," to use his own words. "is there some mystic reason why n% distortion in the loudspeaker doesn't matter but n% cent in the output stage does?"

S. W. AMOS, G. G. JOHNSTONE.

Design for a Pre-amplifier

D. H. W. BUSBY, on p. 328 of his article in your July issue, says that a capacitance of 400pF could be placed across the output with the gain control fully advanced,

with negligible loss of output at 15 kc/s.

This may well be so, but if the gain control is turned down to half-way, the reactance of the capacitor would appear across $50 \,\mathrm{k}\Omega$ and would be fed from a source impedance of at least $50 \,\mathrm{k}\Omega$. In this case the response would be down by at least an extra 3.5 dB at 15 kc/s. In addition, the amount of high-frequency cut introduced would vary with gain control setting, being greatest with the slider electrically at centre, and would become progressively less on either side of this position.

London, S.E.26.

W. C. R. WITHERS. London, S.E.26.

The designer of the pre-amplifier writes.—Mr. Withers is quite correct. The amount of high frequency cut will

indeed vary with the gain control setting. If the loss due to the interconnecting cable is not to exceed 1 dB at 15 kc/s for any setting, the corresponding permissible value of capacity is approximately 150 pF. This will normally correspond to at least 7-8ft of cable, which, for most purposes, will be found adequate.

D. H. W. BUSBY.

 $egin{array}{ll} Damping & Factor: A & New & Approach \end{array}$

AS the one originally responsible for introducing the term "damping factor" the writer feels some responsibility for finding an alternative form now that we are so deeply in the morass. The term had many shortcomings but it could, at least, be used safely so long as it was always finite and positive. The commercial release of amplifiers with negative damping factors has been very confusing to engineers, to say nothing of the general public. For an increase of 22% in total circuit damping, the "damping factor" increases from 10 to infinity, then returns back from —infinity to —10. All these extra-ordinary changes in the damping factor would lead one to believe that something important was happening. In reality nothing has happened except a slight and steady increase in the total damping. The tricks played by the so-called damping factor are due merely to an unfortunate choice of definition. With this definition, instability occurs when the damping factor ≤ -1 .

The total circuit damping is a function of the total circuit resistance, that is, the algebraic sum of the voice coil resistance (always positive) and the amplifier output resistance (positive or negative). I therefore put forward the following as a much more satisfactory and logical substitute for damping factor:

Damping ratio
$$= \frac{R_{\rm L}}{R_{\rm L} + R_0}$$

Where

R_L=load resistance

R₀=output resistance of amplifier

and where both R_L and R₀ are referred to the same side of the transformer. The following table is for $R_L = 15$ ohms and is purely

as an example:

R _o ohms	R _L +R _o ohms	Damping factor = R _L /R ₀	$\begin{array}{c} \text{Damping} \\ \text{ratio} \\ = R_{L}/(R_{L} + \\ R_{0}) \end{array}$
+75 +3 +1.5 +0.15 0 -0.15 -1.5 -5.0 -12.0 -13.6 -14.3 \(-15.0	+90 +18 +16.5 +15.15 +15.0 +14.85 +13.5 +10.0 +3.0 +1.4 +0.7	$\begin{array}{c} +0.2 \\ +5 \\ +10 \\ +100 \\ \infty \\ -100 \\ -10 \\ -3 \\ -1.25 \\ -1.1 \\ -1.05 \\ -1.0 \end{array}$	0.167 0.83 0.91 0.97 1.0 1.01 1.11 1.5 5.0 10.7 21.4
on verge o	f instability	<u> </u>	

It will be seen that the proposed damping ratio is positive and finite so long as instability does not occur. It is also proportional to the actual damping in the circuit. It appears to be the only available function with all the desired qualities.

F. Langford-Smith

Amalgamated Wireless Valve Company, Sydney, N.S.W., Australia.

^{*} Langford-Smith, F., "Radiotron Designer's Handbook", 3rd ed. 1940.

Band II F.M. Tuner Unit

Design Suitable for Use With a Wide Range of A.F. Amplifiers

ALTHOUGH the tuner circuit described in the following pages was designed primarily for use with the Mullard 5-valve 10-watt amplifier circuit¹, or with the 20-watt circuit using EL34s², it is suitable for use with a wide range of amplifiers. The frequency range covers the whole of Band II (87.5-100 Mc/s), and while the circuit design chosen incorporates some of the more modern developments applicable to this type of reception, the construction is kept free from complication. The power supply would normally be taken from the main audio amplifier.

Circuit Description .- A complete circuit of the tuner unit is given in Fig. 1. The r.f. stage is a conventional pentode amplifier circuit, using a Muliard EF85. The associated aerial circuit is pre-tuned to the centre of the Band II range. In the anode circuit is an r.f. choke (L_3) , with the parallel-fed r.f. tuned circuit $(L_4, C_9, C_{10}, C_{11})$, of which C_{10} forms one section of the two-gang tuning capacitor. The r.f. voltage, taken from a tap on L₄, is fed into the grid circuit of a Mullard EF80 operating as a selfoscillating additive mixer. The oscillator section of the mixer is basically of the tuned-anode type, with a tuned circuit consisting of L₆, C₁₂, C₁₃ and C₁₄ (C₁₃ forming the second section of the tuning gang.) The intermediate frequency developed by the mixer valve is fed to two conventional i.f. stages using Mullard EF41s, at an intermediate frequency of 10.7 Mc/s. In the second i.f. stage the EF41 operates as a partial limiter valve at high signal levels, and the bias developed across its associated grid circuit capacitor C_{32} is fed back to the first i.f. valve and to the r.f. valve. The final i.f. stage drives a ratio detector circuit containing a Mullard double

diode, type EB91. Audio-frequency voltages developed in the detector circuit are taken through a 50-microsecond de-emphasis network consisting of R_{16} and C_{45} , and thence to the a.f. output socket for feeding to the audio amplifier.

Aerial Circuit and R.F. Stage.—The aerial circuit has been designed to be matched to a 75- Ω balanced feeder line, thus permitting a simple connection from a conventional type of dipole aerial. In cases where the feeder line of this type is unscreened, L_1 may be centre-tapped to earth so that any noise voltages picked up in the feeder itself are reduced. Dust core tuning is employed, and the resonant frequency of the grid tuned circuit is arranged to be 94 Mc/s. The total tuning capacitance in the grid circuit is of the order of 18pF, of which C_2 forms 5pF and the remainder is formed by the valve input capacitance, plus stray capacitance. The input damping of the valve amounts to 3,800 Ω , giving

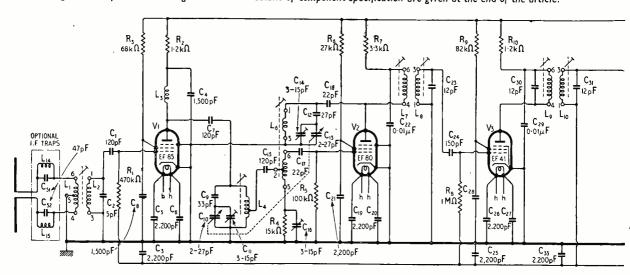
an effective secondary circuit impedance of the order

of $1,600~\Omega$ (without the aerial circuit connected). When attached to an appropriate feeder cable, the

aerial circuit bandwidth is 10.8 Mc/s for 3 dB down on 94 Mc/s and the measured aerial gain is 14 dB. C_{10} which has a maximum capacitance of 27pF tunes the r.f. circuit, and the series capacitor C_9 and trimmer C_{11} are added to track the r.f. circuit correctly to the oscillator circuit. Thus the equivalent capacitance swing is limited to approximately 8pF, and, in addition to the lumped capacitor constants, a further 12pF and 4pF are added to the circuits in the form of the r.f. valve output capacitance plus strays and the equivalent input capacitance of the mixer reflected into the tuned circuit, respectively.

To assist further in obtaining correct tracking over

Fig. 1. Complete circuit diagram. Further details of component specification are given at the end of the article.

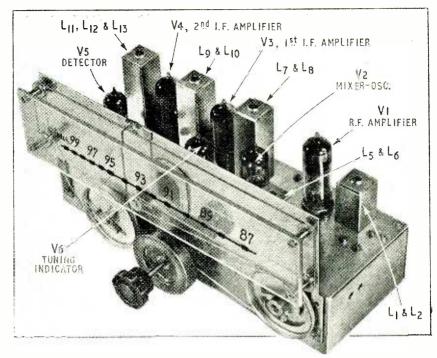


By L. HAMPSON, B.Sc.*

the whole of the band, L4 is tapped by means of C₁₅, so that the loaded Q factor of the r.f. circuit is comparatively high. The average value is 75, and a mean bandwidth of 1.3 Mc/s for 3 dB attenuation is maintained over the whole tuning coverage, for any point in the band. Therefore, with an equivalent load impedance of the order of $4.5k\Omega$ at 94 Mc/s for the r.f. tuned circuit, a theoretical gain of 43 dB is obtained for the r.f. stage (including the aerial circuit). Under the circuit conditions shown the EF85 operates at a mutual conductance in the region of

9.5mA/V. Measurement of the stage gain showed it to be only slightly less than calculated.

Mixer Circuit.—The self-oscillating type of mixer adopted in this circuit, has proved highly popular in countries where f.m. reception is well established. For successful operation the frequency difference between the developed intermediate frequency and the incoming signal frequency should be large. Fortunately this is generally so in f.m. receiver applications, where the intermediate frequency is usually about 10 Mc/s. As its name implies it is essentially an oscillator with provision for feeding in an r.f. signal, so that additive mixing occurs on



Layout of components on the top of the chassis.

a common electrode (in this case the control grid of the EF80). The anode circuit contains an i.f. transformer which is tuned to the intermediate frequency developed.

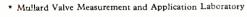
The voltage gain of this single valve circuit is equivalent to that of a two-valve stage consisting of an oscillator and a separate mixer. At the same time the inherently low equivalent noise resistance obtainable with additive mixing is retained. In addition the use of a single valve makes the circuit economically more attractive.

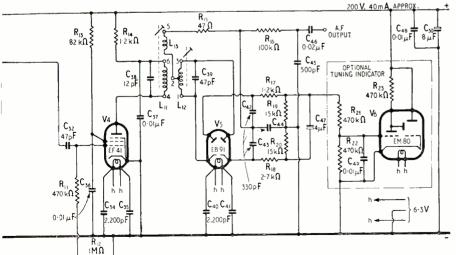
In a mixer of this type, it is generally essential to have some form of "isolation" between the r.f. tuned circuit and the oscillator circuit of the mixer to prevent interaction and pulling of the oscillator section. This is achieved

section. This is achieved by operating the oscillator in a bridge circuit, the equivalent circuit of which is shown in Fig. 2. It will be seen then, that for the bridge circuit to be in balance, the relation,

$$C_{16} = \frac{C_{_{\rm in}} \times L_{_{5A}}}{L_{_{5B}}}$$

must hold. When this condition has been achieved there will be minimum interaction between the two relevant circuits. An obvious added advantage of this bridge connection is that when in balance, there will be minimum oscillator voltage at the r.f. input point. This is important in order





WIRELESS WORLD, AUGUST 1955

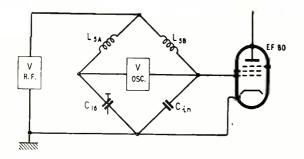


Fig. 2. Equivalent bridge circuit of oscillator section of mixer valve. C_{in} represents the valve input capacitance.

to keep the oscillator voltage at the aerial terminals as low as possible. In this circuit, it is possible to reduce the oscillator voltage to an average value of 100mV at the r.f. input point.

The oscillator circuit is operated at a frequency higher than the signal frequency, i.e. at approximately 97 to 111 Mc/s. As with the r.f. circuit, a series capacitor C₁₂ and a parallel trimmer C₁₄ are included for tracking purposes, giving an effective swing of the tuning capacitor C_{13} of approximately 7 pF. To keep the oscillator circuit as stable as possible the total capacitance associated with the tuned circuit has been made as large as possible, consistent with obtaining sufficient oscillator drive for the mixer valve. Also the cathode of the mixer valve is directly earthed in order to avoid capacitive hum modulation.

The blocking capacitor C₁₈ also forms the tuning capacitance for L_7 , since L_6 is effectively a short-circuit at the intermediate frequency. Similarly L7 forms an r.f. choke at the oscillator frequency. It will be seen then that the effective lumped tuning capacitance

across L, is equivalent to:

$$\frac{C_{22} \times C_{18}}{C_{22} + C_{18}}$$

and the voltage tap down in the i.f. transformer is equal to

 $\frac{C_{\text{out}} + C_{18} + C_{22}}{C_{\text{appairance}}}$ where C_{out} is the valve output capacitance.

However, as the value of C22 is so much higher than C_{18} , the loss in gain of the mixer is negligible.

In order to ensure that the mixer valve operates on the optimum point of the conversion conductance curve, it is recommended that the oscillator grid

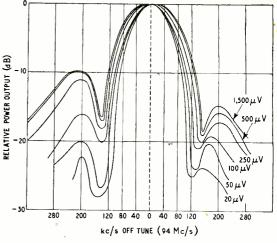


Fig. 3. Tuning characteristics for various input signal levels at the aerial terminals, with a frequency deviation of 22.5 kc/s.

current should not fall below a level of 35 µA in any part of the band. It will be found in operation that the oscillator grid current will not vary by more than approximately $\pm 10\%$ of the mean value over the tuning range.

A conversion conductance of 2.5 mA/V is obtainable with the EF80. Therefore, with the i.f. transformer used, the mixer gain from the tap point of L5 to the signal grid of the first i.f. valve will be 32 dB. Due to the tapping of L₄, the effective gain of the r.f. stage is proportionately reduced to 34 dB. At 94 Mc/s the overall "front end" gain is of the order of 66 dB, measured from the aerial terminal to the control grid of the first i.f. valve.

I.F. Stages.—The design of i.f. stages for f.m. receivers presents a considerable number of conflicting problems. Very briefly summarized they are:

There should be adequate transmission of the significant side currents.

(ii) A good measure of adjacent channel selectivity is required.

(iii) There should be a reasonably linear phase/ frequency characteristic in the transformers.

(iv) A certain allowance in the bandwidth should be made for small random drift in the oscillator.

(v) The uses of comparatively large values of overcoupling to give a wider bandwidth can produce a high degree of amplitude modulation on the carrier wave, and may give rise to ringing with impul-

sive interference.

It was decided that for the two i.f. transformers used in this tuner, a coupling factor K of design centre 1.2, would be most suitable to meet a compromise for the above requirements, provided the average loaded Q factor of the tuned circuits in the i.f. transformers is in the region of 60 to 70. (K = $k\sqrt{Q_pQ_s}$, where k is the coupling coefficient and $Q_{\scriptscriptstyle \rm p}$ and $Q_{\scriptscriptstyle \rm s}$ refer to primary and secondary windings.)

TABLE I

	lst I.F. Trans- former		2nd I.F. Trans- former		Ratio Detector	
	Prim. L ₇	Sec. L ₈	Prim.	Sec. L ₁₀	Prim.	Sec. L ₁₂
Fixed tuning capacitance (pF) Valve capacitance + strays (pF)	22	12	12	12 10	12	47
Loaded Q in circuit	55	55	68	60	35	26
Coupling factor	1.25 15.8		1.25 21.2		0.65	

In addition to the above requirements it is essential that feedback through the anode-to-grid capacitance of the valve, should be kept small. This usually calls for a strict limit on the maximum usable transfer impedance obtained in an i.f. transformer when applied with any particular valve type. With the EF41, the recommended maximum design centre transfer impedance for the i.f. transformer is 21 k Ω at 10.7 Mc/s, taking into account the added effective anode-to-grid capacitance in the valveholder, and assuming identical impedances in grid and anode circuits.

Table 1 gives a summarized performance of the transformers used in this tuner. All the values quoted

were measured in circuit.

The two EF41 valves operate with a mutual conductance of 2.3 mA/V, and from the relevant impedances given in Table 1, it can be calculated that the total i.f. gain, from the control grid of V3 to the anode of V4 will be 64 dB. Actual measurement showed a slightly lower value.

Measurement of the overall bandwidth of the first and second i.f. transformers gave approximately 210 kc/s for 3 dB, and 600 kc/s for 20 dB, attenuation. The response curve is flat for approximately 70 kc/s,

with a slight dip at the centre frequency.

The coils for the i.f. transformers are wound on common formers, details of which are given later in the Appendix. An inherent drawback in this method of construction is that movement of the dust cores can materially alter the coupling factor if the primary and secondary windings are, by necessity, brought too close together. This is to some extent eliminated in the design presented here, by separating about 25% of the total windings on the coils, so as to form small coupling coils at the earthy end of each winding. Thus the dust cores are kept at least 15 mm distant in the main body of the windings and little measurable

difference in the coupling factor is obtained when the transformers are tuned to ± 1 Mc/s of the correct working centre frequency. The unloaded Q factors of the coils are about 90, with the exception of L_{12}

which is approximately 85. As previously mentioned, on high signal levels V4 is driven into appreciable grid current, and the derived bias voltage developed across R_{11} and R_{12} is used as semi-automatic gain control. Positive peaks of amplitude modulation appearing on the carrier wave are therefore clipped in the grid circuit. With V4 primarily designed as an amplifier, a very high input voltage to the grid is required before saturation occurs. Therefore with only partial limiting occurring, the valve does not deal completely with positive—or negative-going amplitude variations of the carrier. For similar reasons partial limiting may give only restricted elimination of impulsive interference. The time constant of the grid circuit of V4 has been set at 25 microseconds to deal with the upper frequency limits of amplitude variations in the carrier. The small changes in the input capacitance of V4 due to limiter action in the grid circuit and of V3 due to the restricted range of applied bias voltage were not found to introduce any serious deterioration in the required performance of the band-pass filters. Further, harmonics produced by limiter action do not usually cause trouble, as they will be tuned out in the anode circuit of V4.

Ratio Detector—A balanced ratio detector circuit incorporating a double diode (V5) is used in this tuner. To give optimum a.m. rejection under working conditions, R₁₇ should be adjusted for each individual circuit and is specified here as a nominal value.

The value of a.m. rejection measured in the circuit was 46 dB at the centre frequency (10.7 Mc/s) for 30V r.m.s. of i.f. voltage at the anode of V4, falling to

View of underside of chassis with some of the more prominent components identified.

28~dB for $\pm75~kc/s$ detuning of the signal. Similarly for 20 and 10V r.m.s. of i.f. signal, the values were 34 and 26 dB respectively, with corresponding values of 22 and 17 dB for \pm 75 kc/s detuning. These figures include limiter action produced by V4. The peak separation of the "S" shaped detector curve is 320 kc/s.

Measurement of the tuning characteristic is advantageous in a prototype f.m. receiver, in order to examine the side responses which are inherent in most f.m. detector systems. These are shown in Fig. 3, where the tuning characteristic of the unit is plotted for various values of input signal.

These characteristics can be regarded as typical for a receiver equipped with a ratio detector circuit. The side responses are shown at about ± 200kc/s from the centre frequency. To a certain extent these side responses are controlled by the selectivity characteristic of the preceding i.f. stages, and also by the action of a.g.c. In general, receivers designed with a rounded-top overall i.f. response curve, and a comparatively wide peak separation in the detector system help to reduce side responses and produce a peak in the audio output, when the signal is in tune. It may be noted in passing that, with the Foster-Seeley detector circuit and limiter valves, the side responses may be of a higher value than the main signal response. Although the side responses show a comparatively high value in the graph at the larger signal levels, they are in fact hardly noticeable when tuning through the signal.

The audio output from the detector is taken through the 50-microsecond de-emphasis network R_{16} and C_{45} , and coupling capacitor C_{46} to the a.f. output socket. When the output from the tuner is fed to an audio amplifier it is recommended that the amplifier input impedance be not less than $500\,\mathrm{k}\,\Omega$. For use with the pre-amplifier³ designed for use with the 20-watt EL34 circuit² and with other pre-amplifiers of similar input impedance and sensitivity it is recommended that a correction circuit (Fig. 4) be used to obtain the required input impedance and attenuation.

Tuning Indicator.—An optional tuning indicator using a Mullard EM80 is fitted. The bias voltage for this valve is derived from the ratio detector circuit.

Overall performance.—The total gain of the prototype tuner from the aerial terminals to the anode of V4 was approximately 130 dB for a small signal at 94 Mc/s. When coupled to a Mullard 5-valve 10-watt amplifier, the average sensitivity over the band for 50 mW output was $1.2 \mu V$ with a signal of 22.5 kc/s deviation. The average input signal over the band for 500 mV audio output is approximately $12 \text{ to } 15 \mu V$.

Oscillator radiation.—The average oscillator voltage measured at the aerial terminals was 350 μ V for the fundamental oscillator frequency and 75 μ V for the 2nd harmonic. The average radiated field strengths over the band are 40 μ V per metre and <15 μ V per metre respectively at a distance of 10 metres from the measuring aerial.

Constructional details.—The accompanying photographs show the main layout of the tuner. A chassis of 16 s.w.g. aluminium, dimensions 10 in \times 3½ in, with 2 in depth, is used. With the exception of the valves, i.f. transformers, aerial and oscillator coils and scale assembly, all components are mounted underneath the chassis.

A balanced heater circuit is used. This enables the tuner to be connected to the centre-tapped heater supply of either of the amplifier circuits previously

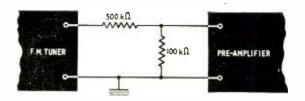


Fig. 4. Simple attenuator recommended for use with amplifiers of high sensitivity and input impedance less than $500k\Omega$.

referred to^{1,2}. Adequate r.f. decoupling of the heater supply lead is essential, in particular to prevent harmonics from the detector from reaching the earlier stages. The ninth harmonic of the i.f. can be particularly troublesome, as this falls in the centre of the tuning range. Strict attention should be paid to the general decoupling of the valve electrodes, and to the tuned circuits. The relevant decoupling capacitors, should be returned to, or very near to, the cathode-to-chassis connection of the valve concerned, with the shortest possible leads. It is also essential that C_{16} should have a short connection to chassis.

The tuner requires an h.t. supply of 200 V at approximately 37 to 40 mA and a heater supply of 6.3 V at 1.6 A. Where the h.t. supply is obtained from an amplifier and exceeds 200 V the necessary dropping resistance should be included in the amplifier itself.

To minimize oscillator drift during the warming up period, C_{12} should be of the negative temperature coefficient type, a component of temperature coefficient 750 parts per million being suitable. This will help to keep the long term oscillator drift to a minimum. In the prototype, C_{14} and C_{16} were formed of a 6.8pF capacitor in parallel with a 1.25-10pF trimmer to make up the nominal total. The tap on the r.f. coil is arranged to be at 0.4 of the total number of turns, counting from the earthy end. Optional i.f. traps L_{14} , C_{51} and L_{15} , C_{52} , have been incorporated in the aerial circuit for use where a high degree of i.f. rejection is considered essential. In most cases they may be found unnecessary.

Neatness in wiring is essential. In particular, in the i.f. stages, components should not be piled over the top of the valveholder as this may lead to an increase of effective anode-to-grid capacitance.

Alignment.—The correct alignment of an f.m. receiver calls for the use of some expensive equipment, but good results can be obtained by using only an a.m. signal generator, covering Band II (87.5-100 Mc/s) and the intermediate frequency of the tuner (10.7 Mc/s). As the first two i.f. transformers are overcoupled, it is essential to damp the transformers whilst they are being tuned, otherwise unsymmetrical response curves may result. A resistor of about $5 \, \mathrm{k} \, \Omega$ is suitable for this purpose and it should be placed across the grid circuit tuned winding, when the anode circuit is being tuned, and vice versa. The resistor can be temporarily held on with a touch of solder.

Connect the signal generator output (10.7 Mc/s) to the control grid of V4 and tune L_{11} for maximum deflection either in the tuning indicator or on a high resistance voltmeter ($20 \, \mathrm{k} \, \Omega/\mathrm{V}$, $10 \, \mathrm{V}$ scale) across C_{47} . Transfer the generator in turn to the grid of V3, and the centre-tap point of L_5 . Tune L_9 , L_{10} and L_7 and L_8 respectively for maximum deflection, using the damping resistor for the coils as before.

To eliminate considerable trial and error in the alignment of the r.f. and oscillator circuit, some approximate values of the correct trimmer settings

are given. C16 can be set initially at about 10 pF, and C₁₄ to 12 pF, with the dust core of L₆ tuning in the base end of the coil away from L₅. C₁₁ is set at approximately 5 pF. The bridge circuit may be balanced by connecting an r.f. valve voltmeter from the tap point of L_4 to earth, and adjusting C_{16} for minimum oscillator voltage. While this operation is being done, the main tuning gang should be set with the vanes about half-way between minimum and maximum capacitance. If an r.f. valve voltmeter is not available, a rough and ready, but quite effective method of obtaining balance is to short-circuit the tap on L4 to earth and observe the change in grid current through R_5 (on $50\mu A$ scale). C_{16} is then adjusted until the change in grid current on shortcircuiting L4 to earth is a minimum.

With the signal generator connected at the aerial terminals and the tuning gang at maximum capacitance apply a signal of 87 Mc/s, and tune L₆ dust core so that a maximum deflection is indicated in the tuning indicator or voltmeter. Re-tune the signal generator to 100 Mc/s and adjust C₁₄ for optimum output with the tuning gang at minimum capacitance. settings may need to be checked a number of times to give the correct frequency range for the oscillator.

Re-set the signal generator to 91 Mc/s and adjust L₁ dust core for maximum output. Adjust C₁₁ for maximum output at 98 Mc/s and finally set L2 dust core for maximum output at 93-94 Mc/s. The dust core of the aerial coil should also be at the base end of the former. To align the i.f. traps apply a comparatively large input signal of 10.7 Mc/s to the aerial terminals, and tune L14 and L15 for minimum indicated output. With the signal generator connected again to V4 grid and with the signal of 10.7 Mc/s adjust L_{12} core, for zero d.c. voltage across C_{44} . This ensures that the ratio detector circuit is reasonably well balanced.

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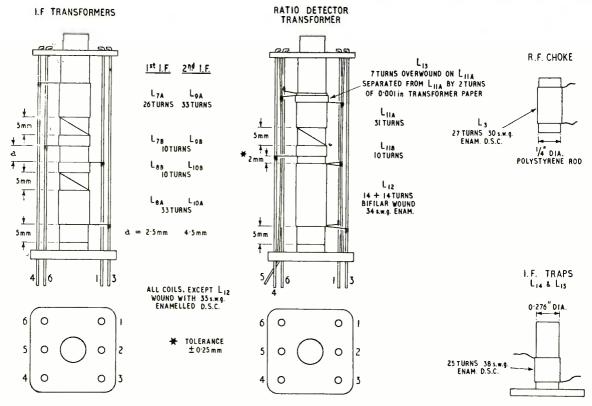
"Design for a 20-watt High-Quality Amplifier, 2—Constructional Details and Performance," by W. A. Ferguson. Wireless World, June 1955.

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(Appendix-Coil Winding Data-on next page)

COMPONENTS LIST FOR F.M. TUNER $C_{34} \\ C_{34} \\ C_{25}$ C_{15} 120 pF \pm 20%(C) Capacitors 2,200 pF±20% 120 pF \pm 20%(C) C₁₆ 3-15 pF (nominal) C_{17} 22 pF \pm 10%(C) $\begin{array}{ccc} C_{6} \\ C_{37} \\ C_{38} \\ C_{39} \\ \end{array}$ 12 pF ± 5% (C or SM) $C_{39} \\ C_{39} \\ C_{39} \\$ 47 pF±5% (C or SM) 5 pF ± 10% (C or SM) $2,200 \text{ pF} \pm 20\%(C)$ C_{18} 22 pF±5%(C) $1,500 \text{ pF} \pm 20\%(\text{C})$ $\binom{C_{19}}{C_{20}}$ 2,200 pF \pm 20%(C) $>2,200 \text{ pF} \pm 20\%(\text{C})$ C_{40} >2,200 pF ±20% (C) C_{21} 2,200 pF \pm 20%(C) $\begin{array}{c} C_{14} \\ C_{14} \\ C_{42} \\ C_{43} \\ C_{44} \end{array}$ 330 pF ±5% $\begin{array}{c} C_{44} \\ C_{44} \\ C_{44} \end{array}$ 500 pF (SM) 120 pF ± 20% C₂₂ 0.01 μF Met. paper C_8 $1,500 \text{ pF} \pm 20\%$ $\begin{array}{cccc} C_9 & 33 \text{ pF} \pm 5\% \\ C_{10} & 2/27 \text{ pF} & Two\text{-gang} \\ C_{13} & (Jackson U101 \text{ S-S}) \end{array} \text{ variable}$ C_{23} 12 pF±5% C_{24} 150 pF $\pm 20\%$ $\binom{C_{25}}{C_{26}}$ 2,200 pF ± 20%(C) 3-15 pF (nominal) (composed of 1.25 —10pF trimmer, C_{11} C_{27} C₄₈ \ 0.01 μF Met. paper C_{28} $0.01~\mu F$ Met. paper C49 (for optional tuning indicator) Wingrove & Rogers, Type C32.01 +6.8 pF, SM C_{29} C_{50}^{-3} 8 μ F 350 V.W. electrolytic 27 pF±5% (optional n.t.c. 750 parts per million) $\left. {\stackrel{C}{C}}_{30} \right. >$ 12 pF \pm 5% (C or SM) C_{51} \rightarrow 47 pF \pm 5%(C) (for optional C_{31} C_{52} \int i.f. traps) C₁₄ 3-15 pF (nominal) C_{32} 47 pF ± 20% C - Ceramic. SM - Silvered mica Resistors (all resistors ½ watt Dubilier "BTS" type) Valves $470 k\Omega \pm 20\% \\ 1,200\Omega \pm 20\%$ $R_{13}~82k\Omega\,\pm\,10\%$ V1 Mullard EF85 R_1 R_2 V2 Mullard EF80 R_{14} 1,200 $\Omega \pm 20\%$ V3, V4 R_3 $68k\Omega \pm 10\%$ $15k\Omega \pm 10\%$ $R_{15}^-~47\Omega\,\pm\,10\,\%$ Mullard EF41 **V**5 Mullard EB91 R_4 $R_{16}~100k\Omega\,\pm\,10\,\%$ $100 \text{k}\Omega \pm 10\%$ $27 \text{k}\Omega \pm 10\%$ $3,300\Omega \pm 20\%$ $1.0 \text{M}\Omega \pm 20\%$ R_5 V6 Mullard EM80 (Optional tuning indicator) R_6 R_7 $\frac{R_{19}}{R_{20}}$ 15k $\Omega \pm 5\%$ R_8 Coils $82k\Omega \pm 10\%$ R_9 $1,200\Omega \pm 20\%$ $470k\Omega \pm 20\%$ $1.0M\Omega \pm 20\%$ $\frac{R_{21}}{R_{22}}$ 470k $\Omega \pm 20\%$ (For optional tuning in- R_{10} Denco 510/AE Aerial transformer L_1 , L_2 Denco 510/RFC R_{22} R.F. choke L₃ R_{23}^2 470k $\Omega \pm 20\%$ dicator) Denco 510/RF R_{12} R.F. coil L₄ Oscillator coil L₅, L₆ 1st i.f. coil L₇, L₈ 2nd i.f. coil L₉, L₁₀ Ratio detector Transformer, Denco 510/OSC Denco 510/IFT.1 Other Components Denco 510/IFT.2 Miniature tag strips-British Moulded Plastics Type A 5556 Stand-off insulators—Wingrove and Rogers. Type TS1-Denco 510/RDT L_{11}, L_{12}, L_{13} 01/1.I.F. traps L₁₄, L₁₅ Denco 510/IFF Scale and drive assembly—Jackson, Type SL15.

APPENDIX-BAND II F.M. TUNER UNIT COIL WINDING DATA

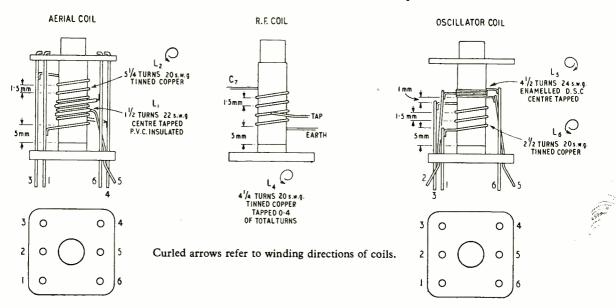


Formers: Aladdin PP No. 5937 (with screening can) or Neosid 5000B.

Former: Neosid type 358/8BA.

Dust Cores: Neosid Grade F 900. Length 16mm. Dia. 6mm. Pitch 1mm.

It is emphasized, that only the absolute minimum amount of sticky tape should be used to hold the i.f. transformer coils in position, otherwise the Q factor will be affected considerably or the self-capacitance of the coils will be increased.



Formers: Aladdin PP No. 5938 (with screening cans) or Neosid 5000A.

Dust Cores: Neosid Grade I 901. Length 12.7mm. Dia. 6mm. Pitch 1mm.

New Navigational Aids

Radio and Radar Equipment at the Paris Air Show

HE twenty-first international aeronautical exhibition, held recently at Le Bourget aerodrome near Paris, demonstrated once again how very closely aviation progress and electronics are allied. Of the 180 exhibitors, representing ten different countries, some twenty were radio and radar manufacturers, and there was equipment from France, Great Britain, the U.S.A., Sweden, Italy and Germany on view.

In the field of radar, the French Compagnie Générale de Télégraphie sans Fil were giving interesting demonstrations of the remote presentation

of radar pictures by a television system. The radar equipment was installed at Pontoise, north-west of Paris, and its images were transmitted by television link to the top of the Eiffel Tower, from which they were retransmitted to Le Bourget aerodrome. The system used depends largely on an analyser storage tube which the C.S.F. have developed and which is described as being a tube capable of storing signals in the form of a pattern of electrical charges deposited on a thin insulating target by a "writing" electron beam. The signals are "read" by a second beam, enabling them to be used to modulate a trans-

mitter. The device is capable of storing the signals for periods ranging from microseconds to hours.

One half of the storage tube can be considered as the radar display system and has deflection coils synchronized with the aerial rotation. Instead of actually showing the image, however, it produces a charge pattern on the target plate in the middle of the tube. The electron beam in the other half of the tube, which has its own scanning system (usually a normal television scan), then discharges the target, the output of which is taken off at a collector electrode and used to modulate the television transmitter. The reading half of the tube is intensity modulated at 20 kc/s in order to avoid interference between the writing and reading scans. Thus on the writing side of the tube the scan can be radial, while on the reading side it can be orthogonal without inconvenience.

The television display equipment appeared to con-

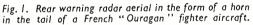
sist of domestic television receivers and they were relaying a radar picture of the Paris air traffic control zone. Until actual pictures are seen it is difficult to realize the full value of the storage tube for air operations, since the inherent tracking feature is so unusual to those who are used to seeing normal persistence radar screens. With the long "memory" the aircraft track is left on the screen from the moment it is first picked up until its goes out of range of the radar.

The chief advantages of the system are the inherent

automatic tracking already mentioned and the great improvement in contrast obtained in the television reproduction of the radar image, which can be made far better than the original picture on the radar screen and clearly readable in daylight.

Another unusual radar system, known as the "3D" (for "three-dimensional"), was exhibited by the French firm Radio Industrie. The equipment, which operates in the 10-cm band, provides range, bearing and height information from one aerial which moves only in the horizontal plane, the usual tilting move-

ment of the reflector for height-finding being elimi-This feature is obtained by the use of a system of scanning (known as a Robinson scan) in which the waveguide feed is moved mechanically so as to produce a change in the vertical direction of the beam from the aerial reflector. In practice the vertical beam can be varied over some 15 degrees and the scan rate is about 800 per minute. This form of height-finding is not new, of course, but in the Radio Industrie radar the disprevious advantages of equipments of this kind are claimed to have been overcome. The side-lobe



suppression is stated to be greatly improved by the use of a specially shaped reflector and the beam width throughout the scan remains constant.

Coming now to airborne radar, an interesting feature of the cloud and collision warning radar shown by the Société Française Radioelectrique was the arrangement for preventing reduction in range by cloud "clutter" when the equipment is being used for collision warning. This takes the form of a series of probes which are introduced at an angle of 45° into the waveguide feeding the reflector. These are stated to make the polarization circular, under which conditions echos from clouds and rain are greatly reduced in amplitude. The equipment, which was reduced in amplitude. The equipment, which was shown installed in a full-sized model of an aircraft, uses a 5-in cathode-ray tube and a scanner reflector about 2ft across.

Also noted was a rear warning radar aerial in the tail unit of a French fighter aircraft (Fig 1).

Outstanding among the radio navigational aids on show was a new instrument landing system by the French company C.S.F. It has the advantage over previous systems of operating on a single frequency and of giving the facilities of azimuth, elevation and distance measurement to the pilot on dial-type instruments. As in other landing aids, overlapping radio beams with different modulation frequencies are used for azimuth and elevation measurement and an interrogator-responder is used for the measurement of distance, the interrogation frequency being between 984 and 996 Mc/s.

The use of a single frequency only is made possible by a form of time-sharing pulse transmission. The cycle of transmission is for an azimuth left-hand beam first, followed by an elevation lower beam and then an azimuth right-hand beam and finally an elevation upper beam. The pulse for each of these tranmissions lasts 1/65 second and each is separated by an interval of 1/360 second. There is then an interval of 1/40 second, during which time the transmitter can send out the response signals to the airborne interrogator, using the distance-measuring facility. The cycle of operations is controlled by a rotary switch revolving 600 times per minute (Fig. 2) which feeds into waveguides and thence to the four aerials. Two of the aerials are for laying down the azimuth beams (known as the "localizer") and are placed upon either side of the runway. The other two are "cheese" aerials providing the elevation beams (known as the "glide-path"), one radiating the upper and the other the lower beam to form a 2½° descent guidance path. The pulses sent out are modulated with a square wave of 20 kc/s for the lefthand "localizer" lobe and 24 kc/s for the right-hand lobe. For the "glide-path" the upper lobe modulation is 34 kc/s and that of the lower lobe 30 kc/s. The aerial for the distance measurement is a "cheese" with separate feeds for transmission and reception.

The airborne equipment consists of two small units and a power supply, together with a crossed-pointer "localizer" and "glide-path" indicator. Distance is

indicated on an edge-scale instrument.

Signals can be received from the "localizer" transmitter at a distance of about 25 miles and the distance-measurement transmitter has a slightly shorter range. The "localizer" accuracy is to within $\pm\frac{1}{2}$ ° at $1\frac{1}{2}$ miles from the transmitter and the distance measuring facility has an accuracy to within 150yd up to 5 miles from the transmitter.

A new "talking beacon" made by the Swedish firm AB Gasaccumulator and intended mainly for fighter aircraft does not require the installation of any special receiver in the aircraft as it operates off the normal

v.h.f. equipment on 100-150 Mc/s.

The beacon depends upon sharply defined beams for its correct operation and these are obtained from two aerial arrays mounted back to back with a screen between them (Fig. 3). This permits forward and backward beams to be radiated without mutual interference. When, for instance, the forward beam is pointing in a northerly direction a transmission by voice of the course to steer to reach it is made, while at the same moment, when the backward beam is pointing south, a voice transmission of the reciprocal bearing is also made from a second transmitter. There is a limitation to the number of voice announcements which can be made, determined by the frequency with which information is needed in practice, and this has been settled at a repetition rate of every 30 seconds when flying on a constant compass heading. At this speed the announcements are made every 20 degrees of beam rotation. Since the aerials are relatively simple there are side lobes present which could give rise to false courses unless precautions were taken to prevent them. The side lobes are therefore masked by a third transmitter which is ten times more powerful than the voice transmitters and which feeds into an "H" type aerial

having a figure-of-eight polar diagram. This aerial is so placed that the wanted beams for the two bearings appear in the crevasses of the masking figure-of-eight polar diagram. Thus all of the side lobes are rendered inaudible by the modulation of the masking transmitter, which is synchronized in time with the voice announcements.

The "Narco Omnigator," made by the American National Aeronautical Corporation and exhibited by the French firm Air Tourist, was of considerable interest, being typical of American radio equipment technique for small aircraft. The instrument, which is compactly designed and only weighs 18lb, combines the functions of v.h.f. transmission on 8 channels, continuously tuned v.h.f. reception over the 108-127 Mc/s band, instrument landing system (ILS), v.h.f. aural range (VAR), v.h.f. omni-directional range

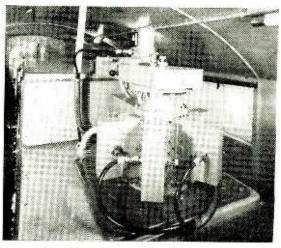
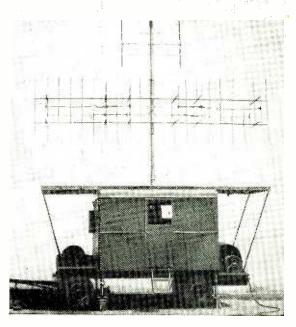


Fig. 2. Rotary switch used at the transmitter of the C.S.F. instrument landing system.

Fig. 3. Swedish "talking beacon" showing the masking aerial mounted above the main array.



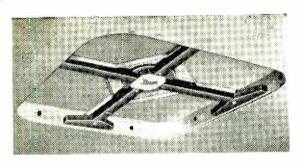


Fig. 4. Bendix radio compass with ferrite-rod aerial elements flush - mounted underneath an aircraft fuselage.

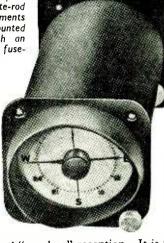


Fig. 5. Cathoderay tube indicator used in the S.I.N.T.R.A. airborne radio combass.

(VOR) and "marker" reception. It is panel mounted and occupies a frontal area of only about $6\frac{1}{2}$ in \times $6\frac{1}{2}$ in.

Flight demonstrations of the latest Decca Navigator flight log computers* were given throughout the period of the show. These devices, which operate from signals provided by Decca Navigator airborne receivers, are very much lighter in weight than previous computers and effect a saving of some 40lb when used in a complete installation. They are entirely electro-mechanical, using no valves, and are designed for operating speeds up to 700 knots. To permit operation at this very high speed, a special memory circuit is included which prevents interference from "lane identification" signal breaks causing loss of "lanes" and is similarly effective during short signal interruptions.

In air navigation nowadays most of the official procedures for approach and departure to and from civil aerodromes are based on radio beacons and ranges. In some flying zones a rapid change-over from one beacon to another is required, and to enable this to be done as quickly as possible three of the French made radio compasses provide means for instantaneous change-over of frequency. In the S.F.R. and Radio Air models a mechanical "memory" arrangement is included whilst in the S.I.N.T.R.A. equipment 18 pre-set, crystal-controlled frequencies are provided. In the mechanically tuned receivers the procedure for enabling the change-over to be made is for the pilot to tune the receiver to the second frequency required -such as the last beacon frequency wanted for a given procedure—and to close a "storing switch," after which he can use the radio compass on any other frequency until the second one is wanted, which he

simply obtains by opening the "storing switch." The time taken for the compass pointer to settle to the new bearing is between 5 and 8 seconds.

In two radio compasses exhibited, including the British Marconi (see last issue p. 306), fixed-coil loop aerials were used, while Bendix showed their new "magnetic antenna." This aerial (Fig. 4) does not entirely dispense with moving parts at the aerial proper but greatly reduces the size of the search coil, thereby increasing the rapidity of action of the direction finder when seeking a new bearing.

The theory of the device is based on the ability of high-permeability ferromagnetic materials to conduct magnetic lines of force easily and to draw into their conducting paths more lines of force than would be found in an equivalent area in free space. In the Bendix loop, four ferrite poles, each with a shortcircuited turn of wire round it, are placed at 90° to one another around a small coil, and their effect is to increase the number of lines of force across the coil. The orientation of the lines of force across the coil depends on the relative signal strengths in the collector rods. Thus pick-up in the quadrants of the aerial causes the maximum concentration of energy across the collector rods which lie transversally to the direction of the transmitting station. The lines of force therefore travel across the loop coil in paths parallel to the collector rods which are receiving the maximum energy. At angles of reception lying between the quadrants, the relative signal amplitudes in the collector rods result in a shifting of the magnetic lines across the coil so that they are parallel to the field in free space.

In the radio compass developed by the French firm S.I.N.T.R.A. the indicator is in the form of a small cathode-ray tube with a graduated scale round its periphery (Fig. 5). This avoids any moving parts in the direction finder proper when used in conjunction with a crossed loop with fixed coils. Operationally the cathode-ray tube indicator has the advantage of immediate indication of the disappearance of the signal and of very rapid indication of transit over the top of a beacon or "coning."

When the loop is installed in the aircraft, one coil is arranged to point in the line of flight, leaving the other at right angles to it. When a signal is received in the line-of-flight loop, its amplitude is proportional to the cosine of the bearing angle of the transmitter being received relative to the heading of the aircraft, whilst the amplitude of the signal received in the other loop is proportional to the sine of the angle. Each loop is arranged to feed into an r.f. transformer and thence into the grids of a double triode valve which also have applied to them an a.f. modulation of equal amplitude but in phase opposition. The outputs from the anodes of the triodes are then fed into a balanced transformer, thereby suppressing the carrier, but leaving the sidebands. The r.f. signal is also received on a small vertical aerial and, after amplification, fed into the r.f. transformer, resulting in a carrier with two sidebands. The phase of the modulation of the carrier is a function of the bearing of the transmitter relative to the heading of the aircraft. The modulated carrier is then amplified and detected and, after suppression of the d.c. component, applied to a pulse generating circuit where a pulse is formed each time the detected signal passes through a maximum. This pulse is applied to the cathode-ray tube and appears as a notch in the circular sweep corresponding to the bearing of the transmitter in relation to the direction of flight.

^{*}See Wireless World, April, 1951, for principles and description of early type.

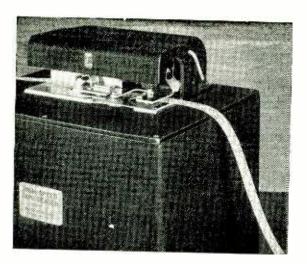


Fig. 1. Coded information in the form of punched holes in paper tape being fed into a high-speed reading device.

A UTOMATION is the magic word that seems to have everyone by the ears (and possibly by the nose as well) in the mass-production industries just now. Radio and electronics engineers, however, appear to be taking the idea very calmly, even with indifference; seeing it perhaps as little more than a new term which has just been invented to describe techniques they have known about for years. And certainly if one looks through the radio and electronics literature for the past decade there is plenty of evidence to support this attitude.

Electronic techniques, of course, are not the whole of automation, but it seems they have a big part to play in the new sphere (in so far as anyone can say what it is). This came out fairly well at the recent conference at Margate on "The Automatic Factory— What Does It Mean?" organized by the Institution of Production Engineers. The titles of some of the papers were in themselves a fair indication—"Automatic Electronic Control of Machine Tools," Atoms, Electrons and Automation," "The Computer—Electronics Contribution to Production," "Automatic Inspection—The Anatomy of Conscious Machines," "Computer-controlled Machine Tools," "The Automatic Office-Industry's Electronic Pulse." Nevertheless it was clear from the content of these papers that the advent of automation (or at least the talk of it) has not so far elicited anything new from electronics. It has merely taken a number of established techniques in electronic measurement, computing and control and placed them under the heading of Automation, together with other equally well-established techniques from mechanical engineering.

It seems, then, that the present conception of "The Automatic Factory" is little more than an ordinary factory to which a large number of electronic (and other) control devices have become attached like barnacles on a ship. This is not greatly significant to the electronics engineer, because it merely means an intensification of his work along the same lines and not a radically new approach. There is, however, a much more highly developed view of the automatic factory which sees the whole organization, and not just the individual bits of machinery, controlled and co-

The Automatic

WHAT SCOPE DOES IT

ordinated by automatic means. (It is sometimes forgotten that there is more to a factory than just the manufacturing processes themselves.) To use a biological illustration, whereas the former conception is equivalent to little more than a mass of disconnected local reflexes all working independently (such as the automatic control of the pupil of the eye with varying light), this highly developed idea is more like a complete animal, with all its sensing and actuating organs not only reacting to local conditions but controlled and co-ordinated by a central nervous system.

Here then is considerable scope for the electronics man—in the design of information-handling systems for overall control and in the linking of these systems to the local electronic control devices. To elaborate on the possibilities we cannot do better than quote from a recent edition of a book* by Norbert Wiener, the American mathematician, who is known principally for his writings on cybernetics. Describing an automobile factory of the future he says: "In the first place, the sequence of operations will be controlled by something like a modern high-speed computing machine. In this book and elsewhere I have often said that the high-speed computing machine is primarily a logical machine, which confronts different propositions with one another and draws some of their consequences..."

"The instructions to such a machine, and here, too, I am speaking of present practice, are given by what we have called a taping (Fig. 1). The orders given the machine may be fed into it by a taping which is completely predetermined. It is also possible that the actual contingencies met in the performance of the machine may be handed over as a basis of further regulation to a new control tape constructed by the machine itself, or to a modification of the old one..."

The computing machine represents the centre of the automatic factory, but it will never be the whole factory. On the one hand, it receives its detailed instructions from elements of the nature of sense organs, such as photo-electric cells, condensers for the reading of the thickness of a web of paper, thermometers, hydrogen-ion-concentration meters, and the general run of apparatus now built by instrument companies for the manual control of industrial processes. These instruments are already built to report electrically at remote stations. All they need to enable them to introduce their information into an automatic highspeed computer is a reading apparatus which will translate position or scale into a pattern of consecutive digits. Such apparatus already exists, and offers no great difficulty, either of principle or of constructional The sense-organ problem is now new, and it is already effectively solved.

"Besides these sense organs, the control system must contain effectors, or components which act on the outer world. Some of these are of a type already familiar, such as valve-turning motors, electric clutches and the like. Some of them will have to be invented,

^{*&}quot;The Human Use of Human Beings," evised edition 1954, Eyre and Spottiswoode.

Factory

OFFER ELECTRONICS?

to duplicate more nearly the functions of the human hand as supplemented by the human eye. . . ."

"Of course, we assume that the instruments which act as sense organs record not only the original state of the work but also the result of all the previous processes. Thus the machine may carry out feedback operations, either those of the simple type now so thoroughly understood, or those involving more complicated processes of discrimination, regulated by the central control as a logical or mathematical system. In other words, the all-over system will correspond to the complete animal with sense organs, effectors, and proprioceptors, and not, as in the ultra-rapid computing machine, to an isolated brain, dependent for its experiences and for its effectiveness on our intervention. . . ."

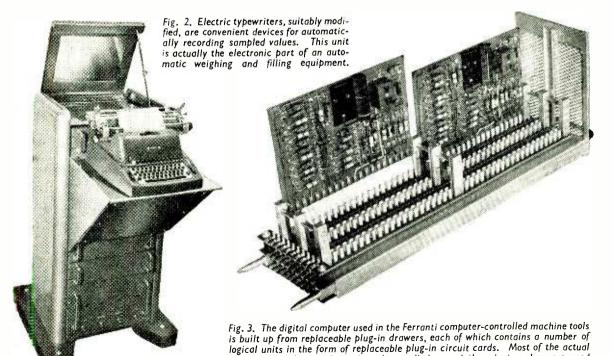
Undoubtedly Wiener's description is very much of a pipe-dream at the moment (or should we say a pipe-nightmare?), but it comes from an informed imagination and should not be dismissed too lightly. Already, in fact, we are beginning to see the initial developments. On the purely manufacturing side there are automatic electronic devices using "sense organs" and "effectors" coming into use for the control of continuous processes, while on the organizational side electronic digital computers, originally introduced into factories for straightforward accounting work, are being used to assemble data for production control purposes. What Wiener refers to as "the general run of apparatus now built by instru-

ment companies for the manual control of industrial processes" could be seen in great variety at the recent British Instrument Industries Exhibition. The newer "information-handling" side was also well represented. For example, both Fielden and Elliott were showing automatic devices for continuously monitoring industrial plant by sampling physical variables (e.g., temperature, flow, level, pressure) at various points and printing out the results on paper (Fig. 2). Electronic discriminating circuits detect when the values are above or below pre-set limits, and cause the appropriate alarms to be given. The British Iron and Steel Research Association were showing how the angular position taken up by a shaft in a selfbalancing servo system can be automatically registered in digital form as a decimal number and a coded version of it recorded as punched holes in a paper tape. Such "analogue-to-digital converters" are, of course, essential components in apparatus for numerical monitoring.

In another type of work an important development for the metal-working industries is the computer-controlled machine tool, which is now emerging from the laboratory and being sold as a commercial product. The Ferranti equipment, which has already been described in Wireless World†, was the subject of a paper by D. T. N. Williamson at the Margate conference, and some of the computer circuitry was shown in a small exhibition which ran concurrently (see Fig. 3). The computer here is used for interpolation between points of change on the contour to be machined, and is a digital machine plotting out the curves point by point. R. H. Booth, of E.M.I. Engineering Development, showed, in another paper, how the same sort of interpolation could be achieved by analogue computing techniques, and there was

† "Electronic Positioning," Wireless World, January, 1955.

computing is done with semi-conductor diodes, and thermionic valves are used



WIRELESS WORLD, AUGUST 1955

only for amplification between logical operations.

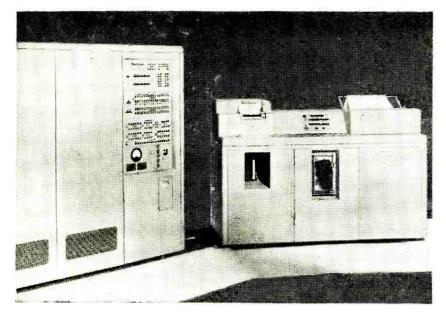


Fig. 4. Commercial electronic digital computer suitable for production planning. The unit on the left contains the programme controller and arithmetic unit, while on the right is a tabulating machine which feeds in information from punched cards (left-hand side) and automatically prints the computed results (right-hand side). This computer and the equipment in Fig. 2 were developed by the British Tabulating Machine Company.

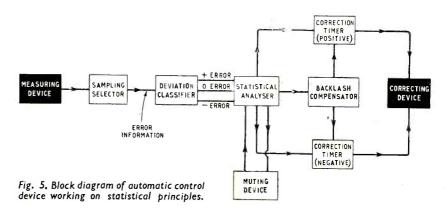
considerable discussion on the relative merits of the two rival methods.

The use of computers for the overall planning of factory work was also discussed by Mr. Booth. He envisaged a machine into which could be fed information on the articles to be manufactured (in terms of the processes needed to make them) and out of which would be obtained information on such things as completion time, costs and requirements for fresh supplies of raw materials. To do this the computer would have to incorporate storage systems containing all the necessary reference data such as stocks of materials, available capacity for machining and assembly and costs of individual processes. A commercially built digital computer of the kind that could be adapted for such work was shown in the exhibition by the British Tabulating Machine Company (see Fig. 4). In the completely automatic factory, however, such a computer would do more than just produce results for human beings to act upon. Also fed into it would be the "operational" information from the plant monitoring systems described above, and this would be assimilated with the "planning" information to produce data from which the computer would control the plant automatically.

A more specialized type of computer for the control of continuous manufacturing processes was described

by J. A. Sargrove and Peter Huggins, of Sargrove Elec-It works on tronics. statistical principles and makes corrections to the processing machinery on the basis of error trends which it detects in the finished product. The apparatus (Fig. 5) includes a mechanical measuring device to sense the variations in the product, a transducer to turn these into electrical signals, a selecting device which samples the product at

suitable intervals and filters out unwanted information, and a "deviation classifier" which quantizes the error information into three distinct categories-" tive error," "no error" and "negative error." A statistical analyser then computes the sum of the errors, retaining in storage their net sense, and actuates correction timers (positive or negative, as appropriate) which by their time of operation control the amount of correction applied to the processing machinery. A "muting circuit," consisting of a timer, disconnects or paralyses the statistical analyser until the corrected product is itself being sampled, while a "backlash compensator" increases the operating time of the timers if there is a change of sense in the correction. Electronic brains" such as these are very specialized and limited in their abilities, with none of the flexibility which characterizes the human brain. It is a good thing, however, that they can be designed to replace the human operator, whose very complexity and desire for better things makes him unreliable as a control mechanism. As a character says in Karel Capek's play about robots, R.U.R .-"anyone who has looked into human anatomy will have seen at once that man is too complicated, and that a good engineer could make him more simply." The electronics engineer is now in the process of learning to do precisely that, although he may not know it.



Wide Range Electrostatic Loudspeakers 3—Com

3—Complete Systems: Loudspeaker/Room Relationships

By P. J. WALKER*

N the first part of this article we showed that for a given size, the apparent efficiency of an electrostatic unit may be increased by reducing the bandwidth which that unit is required to cover. An obvious method of increasing the overall efficiency of a complete electrostatic system, therefore, is to divide the system into a convenient number of frequency bands and to feed them via crossover networks. Optimum design is obtained by increasing gaps and areas with decreasing frequency.

An alternative method of increasing apparent efficiency is to subdivide the loudspeaker area into a number of smaller units each covering the whole frequency range, the units being coupled by inductors so that the whole loudspeaker becomes a transmission line. (Fig. 1.) The acoustic radiation resistance appears as conductance in parallel with each capacitive element. For a fixed total area, and neglecting losses, the efficiency varies directly with the number of

Consideration of these two systems shows that frequency division has considerable advantages over transmission line divisions for most complete systems of domestic size and power requirements. First, if a

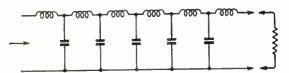


Fig. 1. Capacitive loudspeaker elements coupled with inductances to form a transmission line.

single nine-octave unit is subdivided into a two-unit system, the apparent efficiency is increased 16 times. To obtain the same increase by transmission line division would require a minimum of 12 divisions. Unless the total area of the loudspeaker is large, and the plate separation small, the capacitance of each section of the transmission line becomes very small indeed and requires correspondingly large inductance which must be of relatively high Q.

This apparent efficiency advantage of frequency subdividing over transmission line dividing holds until the bandwidth of each unit is reduced to two octaves.

Apart from transmission line subdivision applied to individual units of a frequency-divided system, practical consideration normally limits transmission line techniques to large-area diaphragms. When such is the case, however, additional facilities are available to the designer both in the accurate control of directional characteristics and in providing a constant phase contour, independent of frequency.

In discussing various possible forms of complete

* Acoustical Manufacturing Co. Ltd.

electrostatic systems, a novel situation arises. The quality criterion of a loudspeaker usually concentrates on three performance parameters, as measured in an unlimited atmosphere. (a) Ability to produce a required sound intensity over the audio spectrum with negligible non-linearity distortion. (b) The sound pressure over the designated listening area should be independent of frequency throughout the audio range. (c) Operation should be aperiodic.

Complete loudspeakers designed on the principles which we have been discussing are capable of meeting these three requirements to a new and exciting degree. We shall see that different designs and approaches differ not so much in terms of (a), (b) and (c) above, but in other factors of importance to quality reproduction; factors which have previously had to take second place or have been masked in the struggle for (a), (b) and (c).

Corner Mounting

There has been a strong tendency in loudspeaker design to make use of the corner of a room. This is because at low frequencies the air load resistance for

a given size of diaphragm is increased 8 times over that of an unlimited atmosphere.

Since the ratio of cabinet "stiffness" to air load resistance is independent of diaphragm size, any increase of resistance due to boundary walls and floors fundamentally reduces the size of cabinet required for a given performance.

As an example, the form of corner electrostatic loudspeaker illustrated in Fig. 2 and designed for full performance down to 40 c/s utilized an internal resonance with a Q of 3 and a built-in enclosure of 10 cu ft. Fundamentally the enclosure size could be reduced either by (1) increasing Q, (2) reducing power and apparent efficiency requirements, or (3)

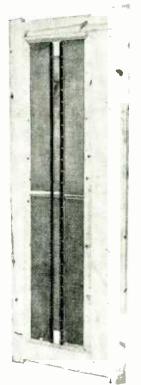


Fig. 2. Wide-range electrostatic loudspeaker in a resonant corner enclosure.

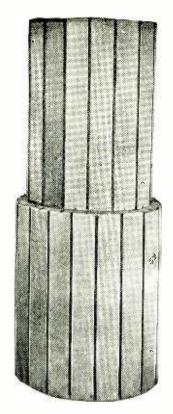


Fig. 3. Cylindrical electrostatic loudspeaker. Each strip carries the full frequency range and the sections are coupled to form an electrical transmission line. The inductor assembly is shown below.



restricting frequency range. Any one factor may be traded for any or all of the others.

It should be noted that with the diaphragm area of Fig. 2 the resistance could be substantially increased by reshaping the whole of the low-frequency area near the floor so that, with the boundary reflections, its dimensions laterally and vertically are similar. Such a form, with a suitably shaped treble unit above it, can be designed to give a level response in direct radiation to the listening area, so that the (a), (b), (c) requirements are not affected. Homogeneity on the other hand, due to the physical spacing of units, is destroyed. This may be more important than is generally realized, particularly in rooms of normal domestic size.

The high-frequency section (centre strip in Fig. 2) is sealed at the rear by an enclosure of width equal to that of the strip and incorporating a fibreglass wedge to offer almost pure resistance throughout the range of the unit. This sealing is necessary in order to maintain front air load resistance by preventing coupling between front and back.

Fig. 3 shows an entirely different form of corner design. The diaphragm area covers the whole surface and extends around the back to form an enclosed cylinder. Every part of the diaphragm carries the whole frequency range. The surface area is divided into units to form a transmission line. The total volume is 15 cu ft. The step in diameter is introduced because the transmission line rotates around the top portion and thence around the bottom portion. The time delay in the sound expanding from the top portion to the diameter of the bottom portion is equal to the time delay of the electrical voltages in the transmission line.

The complete assembly is placed a small distance from the corner of a room so that the boundary reflec-

tors are aiding at the lowest frequency of interest. The large diaphragm area together with the boundary reflections provide a loading approximately equal to ρc at 30-40 c/s. Internally there is acoustic resistance treatment, so that there will be resistive loading at high frequencies, changing to a capacitive load due to the lumped enclosure at low frequencies. Simplified equivalent circuits for high and low frequencies are shown in Fig. 4. The turnover occurs at about 400 c/s and it is obvious that with constant voltage the response will be level above 400 c/s and drop at 6dB/octave below this frequency. This is corrected by progressively rematching to the amplifier below 400 c/s. The section shape may be elliptical to give a degree of direction at high frequencies.

It is obvious that the corner boundaries will introduce peaks and troughs throughout the frequency range. These are, however, exactly the same as occur naturally with live speech or music originating near boundaries in a room. To what degree these effects are important must at the present time be a matter of conjecture. It can safely be said that the subjective effect is by no means as alarming as the appearance

of the response curve.

The advantage of a corner position has already been noted. This advantage is not gained without considerable detriment in other directions. If we wished to excite every room resonance to its fullest extent with a sound source of high internal impedance, we put this source in a corner because this is the position of highest impedance for every mode. In placing our loudspeaker in a corner therefore we are placing it in the worst possible position if our aim is smooth aperiodic sound.

Although the present trend appears to be to tolerate this state of affairs in the interest of the organ's 32ft rank (or reduction of cabinet size), the inherent smoothness of electrostatic loudspeakers once experienced is not lightly thrown away, and there is added impetus in attempts to improve the loudspeaker/room relationship.

Double Wall Enclosure

The strip "twin" unit design of Fig. 2 may be built into a wall in such a way that most room modes are not excited or are excited only feebly. If it is an outside wall, the rear enclosure may be added externally. If an inside wall it may spread over the wall so that from the appearance point of view it has virtually disappeared. Fig. 5 shows the general form of installation. The strip unit extends from floor to ceiling and the low-frequency sections are backed by 5in wide enclosures 4½ft in length, with fibreglass wedges incorporated. The impedances and response are shown in Fig. 5 (June issue). With the dimensions of this example, d=10in since both 5in units are coupled, and the response will be within 3dB of 1 kc/s response down to 35 c/s. These figures include floor, one wall and ceiling, but do not, of course, include the effects of other room boundaries. Assuming a 2in thick wall for rigidity, the volume of a room of 300 sq ft floor area would be reduced by 2%.

There can be no initial excitation of floor to ceiling modes because vertical excitation is evenly distributed. Modes excited in a direction parallel to the wall on which the speaker is mounted will be reduced in number. Assuming a rectangular room, the number of modes excited will be some four times less than

the number excited by a corner floor position.

As can be seen by the following summary, this form of loudspeaker leaves little to be desired.

1. The enclosure being "built-in" can be com-

pletely rigid.

2. The only fold in the enclosure is narrow compared to wavelength and being close to the diaphragm can cause no reflections in the range of that unit.

3. The loudspeaker and its enclosure are completely

predictable.

4. The (a), (b) and (c) requirements previously men-

tioned can be met virtually to perfection.

5. Radiation throughout the whole frequency range is homogeneous; there is no source displacement and no phase problems at crossover.

6. Total radiated energy (as well as axial pressure)

is independent of frequency.

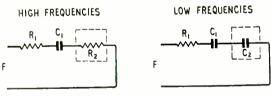
7. The loudspeaker/room relationship is good.

Item 6 deserves further mention. The normal frequency response specification of a loudspeaker is in terms of sound pressure produced on the axis or over a limited listening arc. The mean spherical radiation (total power output) is not usually specified, although it will have a profound effect in a room because the intensity of indirect sound is dependent upon it. If high-frequency radiation is limited to a segment of, say, 90° × 30° (a typical figure) and bass radiation is hemispherical, and if the axis response is level, then there will be a step of 12dB in the mean radiated response. This produces an artificial step in the acoustic ratio (ratio of direct to indirect sound) producing unnatural hardening of the reproduced sound.

Doublet Sources

We now come to consideration of the doublet as a sound source and we shall see that it possesses properties of considerable significance in improving loudspeaker/room relationships. By a doublet we mean a diaphragm, radiating on both sides.

If we assume a 12in-15in unit (moving coil or electrostatic) mounted in a 4ft-5ft baffle, we find that the



R₁ FRONT AIR LOAD R₀ INTERNAL RESISTIVE LOAD

C1 DIAPHRAGM COMPLIANCE C2 ENCLOSURE COMPLIANCE

Fig. 4. Equivalent circuits at high and low frequencies of the acoustic loading on the loudspeaker of Fig. 3.

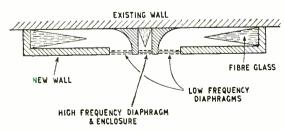


Fig. 5. Sectional plan showing one method of rear enclosure for a strip electrostatic unit.

acoustic system has three main faults. (1) The acoustic air load falls to very low values at wavelengths larger than the baffle size. (2) The acoustic load is very irregular at low frequencies and (3) reflections from the baffle edge occur at higher frequencies. second, and third faults can be mitigated by adopting peculiar shapes.

If, instead of a baffle, we construct a composite electrostatic unit of the same area, the position is completely altered. The resistance per unit area and the total working area are both increased so that the air load is many times that of the baffle case. The load, and consequently the performance, is regular and pre-

dictable.

The construction is that of strip units progressively increasing in plate spacing and area from the centre line. Due to the air load resistances involved for each strip, the permissible bandwidth is reduced over that which could be obtained if the back radiation were sealed off and it is necessary to split the frequency range into three to obtain efficiency comparable to a two-way "sealed" system.

Any unloaded strip considered alone will have a resonant frequency when the diaphragm stiffness reactance equals the air load mass reactance. This however, placed below the frequency range of the strip, so that the mutual radiation of the adjacent strip carrying a lower frequency range increases the radiating area and prevents the application of any effective mass. The complete system is therefore entirely free of resonance except at one low frequency (usually placed at 30-35 c/s). The Q of this resonance is adjusted to maintain response to this frequency.

The complete loudspeaker has a cosine characteristic and this is substantially maintained through the range. It cannot radiate sound in the direction of its surface, horizontally or vertically, so that it cannot excite room modes in two out of the three room dimensions. It will only excite modes in the remaining dimension when placed at a region of maximum velocity for that mode. (The impedance looking into the loudspeaker

is low.)

Having a "cosine" polar characteristic the mean spherical radiation is reduced by a factor of 3 at all frequencies, so that quite apart from freedom of mode excitations any colour due to the room is reduced by a factor of three. This is exactly analogous to a "velocity" microphone. In the same way that a "velocity" microphone is used in place of a "pressure" microphone to reduce studio colour, this "velocity" speaker will reduce colour due to the listening room.

"pressure Listening tests comparing "velocity" speakers of otherwise similar characteristics indicate that a velocity characteristic may well have important features for high-quality reproduction. An electrostatic loudspeaker of this type correctly positioned in the room meets all requirements as did the "wall" form previously described, with the addition of an even better loudspeaker/room relationship. The fact that it requires to be free standing well within the room may or may not be advantageous.

The more the acoustic ratio is reduced (provided always that it is reduced equally at all frequencies), the more one approaches the state of affairs that the pressure at the ears is a replica of the pressure at the position of the microphone in the concert hall or studio (ideal headphone conditions). It must be emphasized that many arguments for and against this condition have been proposed. It is outside the scope

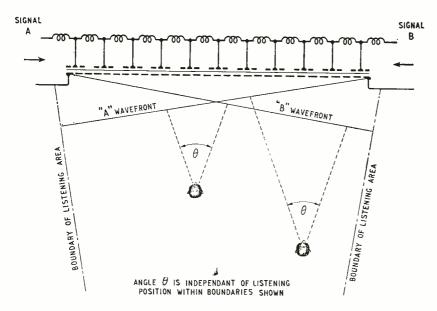


Fig. 6. Stereophony from a single transmission line loudspeaker, with separate channels feeding each end of the line.

of these articles to enter these arguments other than to say that with a monaural channel the choice must be an æsthetic one.

A complete listening room can be designed to produce pressures throughout the room which are more or less equal to the pressures at the studio microphone.

A tube of small diameter compared to wavelength fitted with a piston at one end, and terminated at the other by a resistance of ρc will give pressures anywhere in the tube which are directly proportional to piston velocity and independent of frequency. Provided that the area of the piston equals the tube cross-sectional area, then the requirement of small diameter disappears.

A rectangular room with a diaphragm covering one wall and correct termination on the opposite wall meets the requirements. The space behind the diaphragm must be at least 10in deep and treated like the speaker in Fig. 3. The equivalent circuit is the same as Fig. 4. The sound absorption treatment of the opposite wall must ideally be several feet in depth.

Sound intensity throughout the room is independent of position (including the distance from the diaghragm). The apparent sound source is always in a direction perpendicular to the diaphragm and, of course, moves as the listener moves.

The same loudspeaker may be used for stereophony. With transmission line matching and feeding the signal at one end the wavefront will be tilted, due to time delay. Separate signals may be fed from either end to produce two tilted wavefronts, one for each signal. Since each apparent origin is perpendicular to its wavefront, the aspect angle from the listener is a constant and entirely independent of the listener's position over a large triangular area (Fig. 6). The relative intensity of the two signals is also constant.

A fixed angle, two-channel system of this type may be obtained with a less elaborate listening room. The strip arrangement of Fig. 5 may be installed horizontally instead of vertically. If each unit is a transmission line along its length, then two cylindrical wavefronts will be produced with exactly the

same feature of constant aspect angle already described.

To summarize, the electrostatic principle is capable of surmounting the present limitations of other methods of drive. It is capable of overcoming the present tweeter/woofer concept to produce a closely coupled, integrated assembly. problem of loudspeaker/ room relationships (common to all loudspeakers) still remains, although the design versatility of the electrostatic makes it possible to design for optimum relationship if these can ever be defined for a monaural channel.

A closer understanding of the relative importance of the many factors involved are needed. (a) Source movement with frequency,

(b) Homogeneity, (c) Acoustic ratio, (d) Mode excitation, (e) Phase contour, etc. All are factors which can only be tentatively assessed after long usage.

The author wishes to thank Ferranti Ltd. for permission to publish the result of work jointly carried out, and to acknowledge the invaluable work of D. T. N. Williamson, W. D. Ol iphant, and their colleagues at Edinburgh. Thanks are due to J. Watson, J. Collinson and others at the Acoustical Manufacturing Company, and to the several specialists who have been able to assist with problems in their own field.

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Printed Circuits by the "Plasmet" process. Hints to designers on the preparation of suitable circuit layouts given in a folder from Printed Circuits, Whadcoat Street, London, N.4.

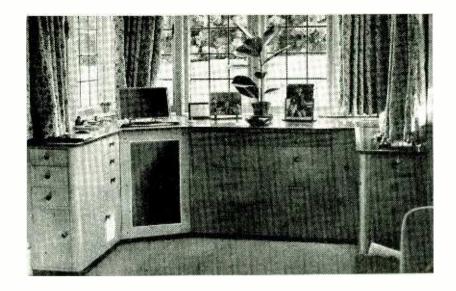
Test and Measuring Instruments by Hewlett-Packard (U.S.A.) obtainable in this country at 8s per dollar plus duty where payable. Short catalogue, giving brief details of major instruments, from Livingston Laboratories, Retcar Street, London, N.19.

Ceramic Trimmer Capacitors suitable for pre-set adjustment of r.f. and oscillator circuits in television tuners. Maximum capacitances from 3 to 20 pF. Also electrolytics with very low leakage currents and subminiature electrolytics for transistor circuits. Technical bulletins from The Telegraph Condenser Co., North Acton, London, W.3.

Band-III Television Convertor with PCC84 s.f. amplifier and ECC81 oscillator and frequency changer. Models available for double-sideband and single-sideband receivers (for London area). Brief technical specification on a leaflet from Invicta Radio, 100, Great Portland Street, London, W.1.

Tape-to-Disc Transfer Service and other recording services Leaflet giving brief details and prices of records from Sound News Productions, 59, Bryanston Street, Marble Arch, London, W.1.

General view of the complete fitment. At the left-hand side of the centre section is the 15in speaker, with the treble speaker assembly on the top of the fit-ment. The only part of the top that opens is on the extreme right-hand side where the transcription motor is situated. The upper part of the centre of the fitment is a steel drawer in which are contained the tape recorder, f.m. tuner, pre-amplifier, amplifier and cross-over unit. Three drawers underneath each accommodate 40 boxes of tape. Five cupboards provide storage space for records or tape and in nine small drawers tools and stationery are kept. The drawer underneath the gramophone turntable was made to correct dimensions to house a ribbon microphone.



High-Fidelity Home By Richard Arbib

A SOLUTION TO THE PROBLEM OF LIVING WITH AUDIO EQUIPMENT

T is hard to determine whether it was the difficulty of finding room for the 91st tape, the wish to have sand in the sitting room, or the fact that the new Garrard 301 transcription motor was so massive that it would not fit along the side of the Leak "Varislope 2" pre-amplifier in the only available space in the existing cabinet. They had both been mounted temporarily in tiers and it was very awkward to adjust the controls on the pre-amplifier by threading the hand under the transcription motor. Incidentally, if it is thought that the possession of 91 tapes is plutocratic, it should be remembered that quite a number of people collect thousands of stamps.

Built-in Furniture Fitment. At last the decision was made. The window alcove in the sitting room should house an elaborate furniture unit to incorporate all the equipment which had been spread around the room or kept in the radio-gramophone cabinet. This had been made specially, more than a quarter of a century ago, with the idea that it would accommodate any developments of the future. However, this thought had now been proved to be erroneous, for it was designed at a time when life was simple and gramophone records were made to play at only one speed, tape machines had not been invented and only one loudspeaker was used at a time in a room.

In the window alcove there had been a writing desk, and at once an argument arose as to how one was to write letters without a desk. However, it was remarked that in this electronic age, one should not write letters at all, and writing was needed only for cheques. Any correspondence could be undertaken on tape and either transcribed by a secretary or heard by the recipient.

As is usually the case in planning any new development, the first thing was to undertake some research and thus many Audio and High Fidelity American magazines were perused. In their pages are illustrated the elaborate high-fidelity installations which, with ranch houses, Cadillacs and typewriters that type as though they are printing, have become the essential acquisitions of the successful American executive. However, an examination of these designs showed clearly that whilst they housed all the pieces of equipment which were considered necessary for the addict of high fidelity, much of the apparatus was situated in a way which was most difficult to operate. It is very nice to have a tape recorder built into a room, but many will agree that it is inconvenient to have to lie on the floor to change the reels.

Basic Principles. In designing this equipment, therefore, it was decided that three basic principles had to be observed. First, everything had to be right technically. For example, the pickup must be some distance away from the loudspeaker. All equipment must be accessible for operation and must be capable of being withdrawn from a fitment in a time not exceeding five minutes. To the service engineer who has to struggle with the servicing of many television sets in a day, this time may appear to be excessive, but in view of the reliability of the units concerned, it was considered that five minutes would not be an undue time. Furthermore, it was also thought that every standard unit should be unaltered. For example, the tape recorder should not be taken out of its original case, for if any section did want servicing, it could be withdrawn as a complete unit and thus returned easily to the manufacturer.

After having attempted for some years to read the titles on tape boxes sideways in bookcases, it was decided that the majority of the tape boxes should be kept in drawers in such a way that they could be located easily. It was thought that the basic storage

space available should be capable of housing 100 12in records and 200 reels of tape.

Although the window alcove was 7ft wide and 5ft deep, one snag was that a height of only 31in was available under the windows. However, it was agreed that the whole of this space should be occupied by the units, loudspeakers and nine small drawers for stationery and odd tools which had been kept previously in the desk. As, apparently, most wives insist on putting vases of flowers, photographs, ash trays and other bric-à-brac on top of furniture, it was arranged that only one small portion of the installation should have a lid which opened. Although record changers may be placed in drawers, it was realized that a transcription motor with a pickup weighing only a few grams should not be mounted in anything movable, and thus of the whole surface area of 22 sq ft only 3 sq ft lifts up.

Accommodating the Units. The basic units which had to be accommodated comprised a 15-in Wharfedale loudspeaker, which, from the counsels of Mr. Briggs, it was appreciated must be housed in a cabinet exceeding 9 cu ft, 8-in and 5-in speakers which should be mounted in an open form of baffle, a Ferrograph Model 2A tape recorder, a Leak TL12 amplifier and "Varislope 2" pre-amplifier, an American Browning f.m./a.m. tuner, which had been brought over to this country long before British f.m. tuners were available, and the Wharfedale 3-way cross-over network unit. Furthermore, a panel was required for switches. The gramophone equipment, for which space had to be found, was a Garrard 301 transcription motor with a Leak moving-coil pickup and arm. Arrangements had to be made for feeding the output to the loudspeaker network, which has been in existence for some time and extends to practically every room in the house.

Owing to the comparatively low height of the top level of the fitment, it was decided that the only way in which to operate the Ferrograph and the amplifier controls conveniently, was to mount everything, except the loudspeakers and the transcription motor, in a drawer. From the accompanying illustrations it will be clear how the units are arranged. The steel drawer was made in two sections and welded together, the Ferrograph drops in on the left-hand side, the TL12 amplifier is in the base of the right-hand side. Mounted above it on a shelf is the Wharfedale crossover network unit to the controls of which extension rods have been fitted so that the potentiometers on the medium and treble cross-over network sections are controlled by knobs mounted on the top panel. The f.m. tuner unit is mounted behind and the little available space in front of that is occupied by a felt-lined recess in which a small crystal microphone is kept and which is connected permanently to the appropriate socket on the Leak pre-amplifier. This unit is mounted at the front of the drawer at the same angle as the control panel of the Ferrograph. A ribbon microphone is used also and this is kept in a special drawer below the gramophone motor.

A potentiometer has been added at the tape input socket of the Leak "Varislope 2" so that the levels from the Ferrograph, the tuner unit and the pickup are all the same. On the left of the "Varislope 2" is a special control panel which has four switches and four jack sockets. Reading from left to right (1) switches the TL12 amplifier to the three loudspeakers through the cross-over unit, (2) switches the amplifier to the extension loudspeaker network, (3) switches the

output of the Leak amplifier, either to the three loud-speakers through the cross-over unit or direct to the 8-in loudspeaker, when it is required to reproduce speech only, and (4) switches off everything at the mains, irrespective of the positions of the switches on the individual units. The jacks below are: (1) connected to the three loudspeakers, (2) connected to the external loudspeaker network, (3) connected to the output of the tuner, and (4), connected to the sound output of the television set which is situated in another part of the same room. The jack sockets are provided for the three loudspeakers and the extension network in case it is wished to feed the output of the Ferrograph direct to these speakers without going through the Leak amplifier. In the same way the tuner socket is used for feeding from the tuner into the Ferrograph. The tuner is also connected permanently to the appropriate socket on the pre-amplifier.

Operation.—One length of shielded cable with screened jack plugs at each end and the various sockets are used as follows. If it is wished to record a radio programme the output is taken from the tuner socket on the small panel and fed into Input 2 of the Ferrograph, the monitoring to the tape machine being undertaken by the gain control on it. A programme can be recorded without it being heard in the room, or if it is wished to be heard in the same room or another room the appropriate loudspeaker switches are thrown and the gain control on the "Varislope 2" adjusted to a suitable volume. If it is desired to record a television programme, the jack is plugged into the TV socket on the sub-panel. If the TV programme is desired to be heard through the Leak amplifier either in the same room or around the house, the output is fed from the TV socket to the input jack socket on the pre-amplifier. To record speech from the room the main control switch on the pre-amplifier is turned to "microphone" and the output from the pre-amplifier connected to Input 2 of the Ferrograph. If it is desired to reproduce a gramophone record on tape, the same procedure is observed with the Leak main control knob in the position to suit the frequency response of the record concerned.

Loudspeaker Mounting.—The arrangement for the loudspeakers is that the section of the fitment for the 15-in speaker has the sides made of 1-in wood, 1-in of sand and ½-in of plywood. The front has a port 12-in × 6-in. On the main top of the fitment is situated the assembly for the 8-in and 5-in speakers. As the height of the alcove is only 7ft the two speakers do not point upwards but are at an angle of 45° into the room. In order to obtain a simple and attractive design, they are mounted on a flat baffle which is held by skeleton woodwork at an angle of 45°, the front and the sides being covered by expanded metal, and the back is open.

The Garrard 301 transcription motor is mounted on a motor board which floats on springs. Also mounted on the motor board is a light which illuminates the turntable and an extra switch which stops the motor without any danger of the pickup being moved.

Cabinet Work.—The finish of the whole cabinet work is bird's-eye maple for the exterior of the front and interior of the cupboard doors, motorboard and panel above the tuner unit. The top is of walnut and the drawers are made of mahogany veneered on the outside with bird's-eye maple.

The front of the steel drawer which accommodates the recorder, amplifier, f.m. tuner, etc., is a dummy to the extent that it has five handles or knobs but opens as one drawer. The units in this drawer have, with the drawer, a combined weight of about 2cwt. The drawer is mounted on Admiralty pattern ball-bearing slides and moves forwards and backwards easily. The wood front is hinged at the top so that the recorder and pre-amplifier controls may be operated without pulling out the drawer.

Having all these units in a comparatively confined space does, of course, raise the problem of ventilation. This has been solved by having the underside of the left-hand section of the drawer completely open except for two channels on which the Ferrograph rubber feet rest. The sides of this section of the drawer are cut away to coincide with the loudspeaker fret and ventilation holes of the recorder case. The other side of the steel drawer has slots $17\text{in} \times 2\text{in}$ cut at the lower ends of the back and sides. Holes are provided at the top of one side, in the felt-lined recess, and an expanded metal grille is mounted between the preamplifier control panel and the switch-jack panel, which have purposely been fixed half-an-inch apart.

The large drawer at the side and the two underneath the steel drawer are of the exact dimensions to take boxes of 7in and $8\frac{1}{4}$ in reels of tape respectively. Each drawer takes two rows of 20 tapes which are filed with the titles uppermost and in line with the front of the drawer. Three shelves in the large right-hand cupboard each accommodate 25 tapes. Four smaller cupboards, whilst being of the correct dimensions to accommodate 12in records could, of course, be used for the storage of tape.

Having spent many hours in making the necessary drawings for the furniture fitment and undertaking all the wiring and connections, of course the question asked most frequently by other high-fidelity enthusiasts is, what alterations would be made if it was being designed all over again. The answer, quite honestly, is none. There is only one snag to the whole system which was apparent when it was planned originally, and that is the desirability of having the main loud-speaker assembly on the side of the room opposite the

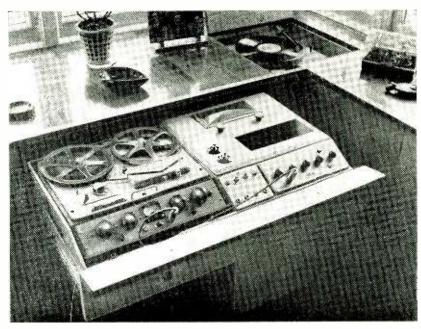
control equipment, so that the volume could be adjusted from a distance. With the space available this was not possible, and the only refinement which could be introduced to the system other than stereophonic sound, for which space is available, would be remote operation for the main gain control of the amplifier.

What Next?—Other friends have said, "Well, now you have everything, what can you do in the future, other than record and listen?" Apart from stereophonic sound which will, of course, be coming along later this year, work has already started upon rewiring the rest of the house with a system whereby the loudspeakers in each room will have individual volume controls and be operated through fourway switches. By this system the occupants of each room and the garden will have choice of the tape or record programmes from the equipment in the sitting room or the Home, Light or Third B.B.C. f.m. programmes. This system is a development of an idea suggested by Douglas Lyons, of the Trix Electrical Co., Ltd. In one cabinet in another room three f.m. tuners, each tuned to a different programme, will be connected to three 10-watt amplifiers. The outputs of these amplifiers will be connected to three pairs of the four-pair cable. The remaining pair replacing the existing extension speaker pair on the main sitting room equipment.

The three f.m. tuners and amplifiers will be switched on and off by a time clock set to switch on at 7 a.m. and off at midnight.

Acknowledgements. — In conclusion, grateful acknowledgement must be made for the assistance of the following friends who have given much technical advice which has been the means of avoiding many of the pitfalls which fall to the experimenter who tries to put together the parts of different manufacturers. Reference is made particularly to Hector Slade of Garrard Engineering & Manufacturing Co., Ltd., Harold Leak of H. J. Leak & Co., Gilbert Briggs of Wharfedale Wireless Works, Ltd., and E. H. Niblett of Wright & Weaire, Ltd. Yes, this article was dictated on tape!

The steel drawer pulled out with the flap covering the controls turned down. The tape recorder is on the left with a tape splicer mounted below the main control knob. Set at the same angle as the recorder control panel is the switch panel and pre-amplifier. On the banel above are the knobs controlling the potentiometers of the cross-over unit and the dial and controls of the f.m. tuner. in the felt-lined recess a crystal microphone is kept and is per-manently connected to the preamplifier. The flap at the top right-hand corner is up to disclose the transcription motor and moving coil pickup. A lamp to illuminate the turntable is fitted on the motor board.



Transistor Equivalent Circuits

2.—Earthed-Emitter Junction Transistors

By W. T. COCKING, M.I.E.E.

N Part 1, we developed an equivalent circuit for the thermionic valve. In doing this we started with the experimental evidence of the characteristics of the valve in the form of a family of curves relating anode voltage to anode current for a series of values of grid voltage. We then approximated these accurate curves by a series of equally-spaced parallel straight lines, which we may conveniently call a linear approximation. The next step was to find an equation which would represent this approximation algebraically and, finally, we found an arrangement of components which could be represented by the same equation. This formed an equivalent circuit.

The accuracy with which an equivalent circuit represents a valve depends entirely upon the goodness of the linear approximation. Over any range of voltages and currents where the straight lines fit the

valve curves closely, the accuracy is good.

With the transistor, the procedure is exactly the same as with the valve. There is a complication, however. It is an experimental fact that all the electrodes of a transistor pass current, not two of them only as in the negative-grid valve. The transistor is more like a positive-grid valve in this respect. Because of this, two families of characteristic curves are needed to describe it and the equivalent circuit is more complex than that of the valve.

The newcomer to the transistor is apt to be misled by certain conventions which are commonly adopted. Most published characteristics are for the transistor in the earthed-base condition and many circuit diagrams show it used in this way. Also, in the usual transistor symbol, the base is represented by a heavy line which is not unlike the cathode symbol of a valve. It is hardly surprising, therefore, that the newcomer gets the impression that the base of a transistor corresponds to the cathode of a valve.

The base is, of course, the equivalent of the grid and the earthed-base transistor is analogous to the earthed-grid valve circuit. To anyone accustomed to valve circuits, it seems very wrong-headed to make the earthed-base circuit the basic one of transistor circuit theory. Actually, there are reasons why it is more suitable than the earthed-emitter circuit, but this is not the stage at which they can well be appreciated.

To the newcomer, however, the earthed-emitter circuit is the natural approach and this is the one that we shall adopt here. At first, only this circuit will be considered and only the junction transistor. This exists in two forms, the n-p-n and the p-n-p. It consists essentially of two pieces of n or p type impurity germanium with a thin slice of the opposite kind sandwiched between them.

One of the end pieces functions as a source of

SUMMARY: As a sequel to the derivation of the equivalent circuit of the thermionic valve in Part 1, equivalent circuits for the n-p-n transistor in the earthed-emitter circuit are developed here. They are based upon the same form of straight-line approximation to the transistor characteristics as was used for the valve. The characteristics of the transistor are represented by four quantities which are easily derived from the characteristic curves; they are base and collector a.c. resistances and forward and backward current-amplification factors.

It is shown that, within the limits of accuracy imposed by the straight-line representation of valve and transistor characteristics, the transistor is exactly equivalent to a thermionic valve having a cathode feedback resistance and a second resistance between grid and cathode.

charge carriers and the other as a collector of them; the one is called the emitter and the other the collector. The meat in the sandwich functions as a control electrode and is called the base. Emitter, base and collector are analogous to the cathode, grid and anode of a thermionic valve. With the *n-p-n* transistor the analogy is rather close, because the internal conduction is mainly by electrons, with the result that base and collector are normally maintained positive to the emitter and the conventional currents enter at base and collector and leave at the emitter.

The *p-n-p* transistor, however, is rather different. Internal conduction is by positive "holes." Base and collector must normally be negative to the emitter and the current enters at the emitter and leaves by the base and collector. There is no valve which behaves like this but if it were possible to make one with a "cathode" which emitted positrons instead of electrons it would do so.

The circuits of typical n-p-n and p-n-p transistors in the earthed-emitter connection are shown at (a)

and (b) respectively of Fig. 1.

The n-p-n Transistor

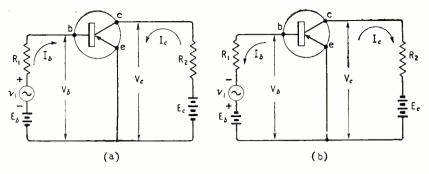
In Fig. 2 are shown the general forms of the characteristics of an n-p-n transistor. The family of curves at (a) relates collector voltage and current for a series of values of base current, while the family at (b) relates base current and voltage for a series of values of collector current. If the curves at (a) were for values of base voltage V_b instead of base current I_b , they would be of exactly the same form as an ordinary set of valve curves and the ordinary equivalent circuit could be used.

Transistor characteristics can be plotted in terms of V_b , but it is customary to use I_b instead, mainly because a transistor has such a low input resistance that we are often more interested in current than in

voltage.

In Fig. 2 (a), the dotted lines represent a linear approximation to the real characteristics upon which we base our equivalent circuit. We proceed in the

Fig. 1. The circuit diagrams of a simple transistor amplifier are shown at (a) for an n-p-n and at (b) for a p-n-p transistor.



same way as in the case of the valve, which was described in detail in Part 1. Some line, not shown, passing through the origin, represents a resistance of value $\rho_{22}=V_c/I_c$. The slope resistance $\delta V_c/\delta I_c$ has the same value for this line. For any other line the value V_c/I_c does not apply but, since the lines are all parallel, they all have the same slope resistance which is, therefore, defined as

$$\rho_{22} = \delta \mathbf{V}_c / \delta \mathbf{I}_c \qquad \dots \qquad \dots \qquad \dots \qquad \dots \qquad \dots$$

We are, for the present, using the Greek letter rho to represent resistance instead of the more usual r, because the latter is commonly used in equivalent circuits derived for the earthed-base connection and we want to avoid any confusion between the two. We are using the subscripts "22" to denote what, by analogy with the valve, we may call the collector a.c. resistance, because that is customary with transistors. This resistance, which we call ρ_{22} , is the transistor equivalent of the anode a.c. resistance of a valve.

The particular line in Fig. 2(a) for zero base current

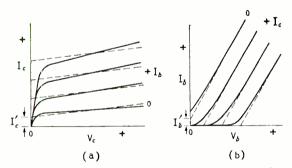


Fig. 2. Collector (a) and base (b) characteristic curves for an n-p-n transistor. The dotted lines represent ideal straight-line approximation to the characteristics.

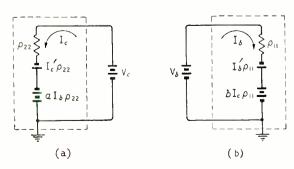


Fig. 3. D.C. equivalent circuit of an n-p-n transistor. The output side is shown at (a) and the input at (b).

 $(I_b = 0)$ cuts the current axis at some current I'_{c} The equation for the line through the origin is

 $I_c = V_c/\rho_{22}$

The equation for the line $I_b = 0$ becomes simply $\mathbf{I}_c = \mathbf{I'}_c + \mathbf{\check{V}}_c/\rho_{22}$

and this derivation is exactly analogous to that for a pentode valve, Fig. 9 of Part 1.

We now define a current amplification factor

for constant V_c. This is merely the ratio of a change of collector current to the change of base current which causes it. The complete equation for Fig. 2(a) now becomes

now becomes
$$I_c = I'_c + aI_b + V_c/\rho_{22}$$
 which is more conveniently written as
$$V_c = I_c\rho_{22} - I'_c\rho_{22} - aI_b\rho_{22}$$

Now what does this represent? On the left-hand side, V_c is the externally applied collector-emitter voltage and is not part of the equivalent circuit of the transistor itself. $I_c\rho_{22}$ is an internal voltage drop due to the current I_c flowing through a resistance ρ_{22} . $I'_{o}\rho_{22}$ and $aI_{b}\rho_{22}$ can be taken to represent e.m.fs acting round the circuit to assist V_{o} but internal to the transistor. The first term of the pair accounts for the offsetting of the $I_b=0$ line of Fig. 2(a) from the origin; the second accounts for the effect of base current upon collector current.

The complete circuit which the equation represents is thus one like Fig. 3(a) in which the transistor part is shown boxed. It is the same as a valve equivalent save for the labelling of the elements.

Now consider the base characteristics of Fig. 2(b). A line through the origin represents a resistance $\rho_{11} = V_b/I_b$. This has a slope resistance $\delta V_b/\delta I_b$ which is the same for all lines. We, therefore, define the base a.c. resistance as

$$\rho_{11} = \delta \mathbf{V}_b / \delta \mathbf{I}_b \qquad .. \qquad .. \qquad .. \qquad (3)$$
 for constant \mathbf{I}_c .

The equation for the $I_c=0$ line is $I_b=I_b'+V_b/\rho_{11}$ We now define a reverse current amplification factor as

$$b = -\delta I_1/\delta I_2 \qquad ... \qquad .. \qquad .. \qquad (4)$$

 $b = -\delta I_b/\delta I_c$ (4) for constant V_b . This is the ratio of a change of base current to the change of collector current responsible for it. The negative sign comes in because it is convenient to have b as a positive number and $\delta \mathbf{I}_b/\delta \mathbf{I}_c$ is itself negative since an increase of Ic reduces Ib. The complete equation is now

The complete equation is now
$$I_b = I'_b - bI_c + V_b/\rho_{11}$$
. This is conveniently written as

 $V_b = I_b \rho_{11} - I'_b \rho_{11} + b I_c \rho_{11}$

This has the same interpretation as the one for the collector circuit and so a circuit to which this applies

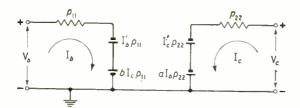


Fig. 4. Complete d.c. equivalent circuit of an n-p-n transistor. For a p-n-p transistor all voltages and currents are reversed.

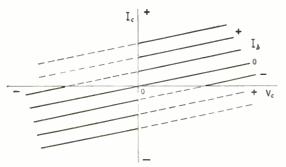


Fig. 5. This diagram as a whole is a graph of a resistance in series with a battery. The part in the first quadrant approximates an n-p-n transistor, while the part in the third quadrant approximates a p-n-p transistor.

has the form shown in Fig. 3(b). However, since $bI_c\rho_{11}$ has the opposite sign to $aI_b\rho_{22}$ it is connected the other way round, to oppose the external driving voltage.

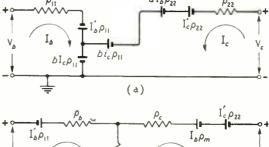
The two circuits of Fig. 3 represent the input and output parts of the transistor and together, as in Fig. 4, they form one equivalent circuit of the *n-p-n* transistor. It is valid for d.c. conditions and accurate in so far as the linear approximations to the transistor characteristics are accurate.

The p-n-p Transistor

We have now to consider the *p-n-p* transistor. It is an experimental fact that its characteristics are of the same form as those of the *n-p-n* but with all the positive signs changed to negative. Because of this, the characteristics are usually drawn upside down compared with Fig. 2. Because *all* the signs for voltage and currents are reversed, it follows that the same equations apply with the signs of all voltages and currents reversed, and so the equivalent circuit of Fig. 4 also applies to a *p-n-p* transistor if all voltages and directions of current flow are reversed.

It should be noted that the reversal of signs applies only to the currents and voltages. The quantities ρ_{11} , ρ_{22} a and b still remain positive. Each of these terms is the quotient of a voltage by a current or the ratio of two currents and the quotient or ratio of two negative quantities is positive.

All this may be a little clearer from Fig. 5. Viewed as a whole, the diagram represents a resistance in series with a battery. The line $I_b = 0$ is for a resistance alone and is merely the graphical representation of a simple resistance. Inserting a battery in series with it to aid the external voltage V_c shifts the line upwards; inserting it to oppose the external voltage shifts it downwards.



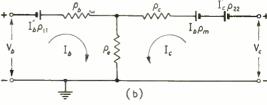


Fig. 6. This diagram illustrates the transformation of Fig. 4 into a different equivalent form. At (a) bl_c ρ_{11} is made common to both base and collector circuits, but its effect in the latter is nullified by the extra battery of the same value opposing it. In (b) this common battery is replaced by a resistance ρ_e , the voltage being developed across it by the current through it.

The characteristics shown by full lines in the top right-hand quadrant are by themselves those of an ideal *n-p-n* transistor. Those in the bottom left-hand corner are those of an ideal *p-n-p* transistor. The dotted parts are non-existent for transistors. The whole diagram can be expressed by one equation which, using the transistor symbols, is

$$\mathbf{I}_c = a\mathbf{I}_b + \mathbf{V}_c/\rho_{22}$$

For the n-p-n transistor V_c is positive and I_c is positive and only the top right-hand part of the diagram exists. Theoretically, I_b can then be negative as long as $aI_b < V_c/\rho_{22}$, but usually I_b is positive. For a p-n-p transistor, V_c and I_c are negative and only the bottom left-hand part of the diagram exists. Again, I_b can be of opposite sign and so positive, but it is usually negative.

One can see, however, quite clearly that ρ_{22} and a must be positive with both types of transistor for, with parallel straight lines, the values of these quantities are independent of the part of the diagram at which they are taken.

Reverting now to the equivalent circuit of Fig. 4, it is possible to transform this to a different arrangement of its parts which is sometimes convenient. The first step in doing this is shown in Fig. 6(a). The only changes that we have made here are to move the positive terminal of $aI_b\rho_{22}$ from earth to the junction of $I'_b\rho_{11}$ and $bI_c\rho_{11}$ and to insert another battery $bI_c\rho_{11}$ in series with it. These make no difference, for the two batteries $bI_c\rho_{11}$ oppose each other in the collector circuit, so the positive terminal of $aI_b\rho_{22}$ is still at earth potential.

The next step is to replace the $bI_c\rho_{11}$ battery common to both circuits by a resistance and to arrange matters so that the voltage drop produced by the currents I_b and I_c produces the necessary e.m.f. The circuit then takes the form shown in Fig. 6(b). If (a) and (b) are to be identical, then all similar voltages and currents must always be the same for each. The conditions for identity are then easily found by writing the mesh equations and equating the coefficients of the same currents and voltages.

For Fig. 6(a) (or Fig. 4), we have for the first mesh

$$V_b=I_b
ho_{11}-I'_b
ho_{11}+bI_c
ho_{11}$$
 and for Fig. 6(b) we have

$$\mathbf{V}_b = \mathbf{I}_b \left(\mathbf{\rho}_b + \mathbf{\rho}_e \right) - \mathbf{I'}_b \mathbf{\rho}_{11} + \mathbf{I}_c \mathbf{\rho}_e$$

Therefore, $\rho_{11} = \rho_b + \rho_e$ and $b\rho_{11} = \rho_e$.

For the second mesh we have for Fig. 6(a) (or Fig. 4)

$${\bf V}_c={\bf I}_{vP22}-{\bf I'}_c\rho_{22}-a{\bf I}_b\rho_{22}$$
 while for Fig. 6(b) we have

$$\mathbf{V}_{\epsilon} = \mathbf{I}_{c} \left(
ho_{c} +
ho_{e}
ight) - \mathbf{I'}_{c}
ho_{22} - \mathbf{I}_{b}
ho_{m} + \mathbf{I}_{b}
ho_{e}$$

hence
$$\rho_{22} = \rho_c + \rho_e$$
 and $a\rho_{22} = \rho_m - \rho_e$

Collecting these results, we find that for identity we must have

$$ho_b =
ho_{11} -
ho_e; \quad
ho_e = b
ho_{11}; \
ho_c =
ho_{22} -
ho_e; \quad
ho_m = a
ho_{22} +
ho_e$$

The resistance ρ_m may be a little confusing at first since it does not appear in Fig. 6 as a resistance. It is, perhaps, better to regard ρ_m as a multiplier for I_b which has the dimensions of a resistance and which when multiplied by I, gives the magnitude of the internal e.m.f. of the transistor.

The circuit of Fig. 6, like that of Fig. 4, represents the static characteristics of an n-p-n transistor in so far as the linear approximation to those characteristics is valid. The circuit for a p-n-p transistor is the same, but with the polarities of all voltages and the directions of all currents reversed.

The usual equivalent circuit is for a.c. conditions only and we could have derived this directly. However, some confusion between the two kinds of junction transistor is avoided if the d..c equivalent circuit is first derived as a stepping stone, and we shall actually find a use later on for the d.c. circuit.

The first step in producing an a.c. equivalent circuit is to produce one which is valid for both d.c. and a.c. We can then get the a.c. one merely by taking away the d.c. part.

Hitherto, the circuits have been, strictly, d.c. ones and we obtain the combined a.c. and d.c. equivalents exactly as we did for the valve by letting each voltage or current be equal to the sum of a mean d.c. component and an a.c. component. Using the sub-

script "m" to denote this mean component and a small letter for the a.c. component, a voltage, say V_b , will become $V_{bm} + v_b$. In Fig. 4, therefore, we replace V_b by $V_{bm} + v_b$, I_b by $I_{bm} + i_b$, V_c by $V_{cm} + v_c$ and I_c by $I_{cm} + i_c$. The currents I'_b and I'_c are pure d.c. quantities and are not affected.

We can now draw the complete equivalent circuits for Fig. 1 as in Fig. 7 and these are valid for both d.c. and a.c. conditions within the limits imposed by the linear approximation. Notice that in both Figs. 1 and 7 the input voltage v_1 is of opposite polarity with respect to earth for the p-n-p transistor (b) compared with the n-p-n transistor (a). This is done for simplicity, so that the positive half-cycle of input voltage in both cases acts to assist E_b .

Exactly as we did with the valve, we can now drop all the d.c. terms from Fig. 7 to leave only the a.c. ones, and we then get Fig. 8. It is very important to notice that the two circuits (a) and (b) are essentially identical. All the voltage generators of (a) are reversed in polarity in (a) compared with (b), and so are all the directions of the currents. If, therefore, we reverse the polarity of v_1 in (b) so that it is positive

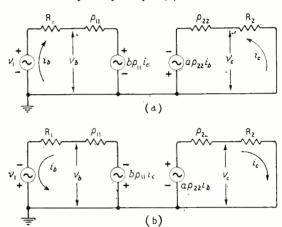
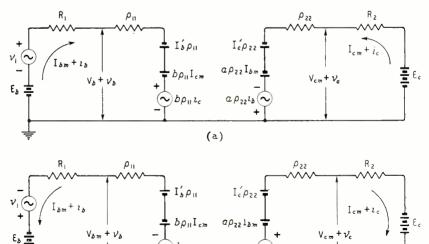


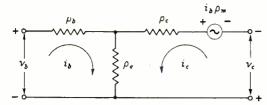
Fig. 8. The a.c. equivalent circuits of Fig. 1 are shown here. Since all voltages and currents in (b) are reversed as compared with (a), the two circuits are identical.



(b)

Fig. 7. The full a.c. and d.c. equivalent circuits of Fig. 1 are shown here for (a) an n-p-n transistor and (b) a p-n-p type. The a.c. inputs are of opposite polarity in the two.

WIRELESS WORLD, AUGUST 1955



An alternative equivalent circuit to that of Fig. 8 is shown here. This is the one most often used for the transistor.

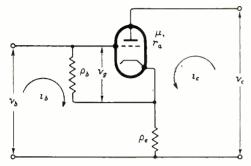


Fig. 10. A valve circuit which has an equivalent circuit the same as that of Fig. 8 and which, therefore, represents a transistor.

to earth as it is in (a), the direction of i_b must be reversed and this will reverse the polarity of the generator $a \rho_{22} i_b$ and hence the direction of i_c . turn, the polarity of the generator $b \rho_{11} i_e$ will reverse.

It follows that the a.c. equivalent circuit is exactly the same for both n-p-n and p-n-p transistors.

This equivalent circuit can actually be used to represent, not only junction transistors, but any device whose characteristics can be expressed by two families of curves similar to Fig. 2. It can, therefore, also be used for the point-contact transistor (although it is not always easy to obtain these curves for this device) and for a positive-grid thermionic valve. For a negative-grid valve the grid current is zero, so ρ_{11} is infinite and b is zero; the left-hand half of Fig. 6 then disappears. The right-hand half forms the usual valve equivalent circuit if μv_g replaces $ai_b \rho_{22}$.

There are a great many alternatives to Fig. 8. Some of them are just ordinary circuit transformations of Fig. 8; others are similar but depend on other quantities than ρ_{11} , ρ_{22} , a and b for defining the transistor characteristics. A particularly common alternative to Fig. 8 is shown in Fig. 9; it is the a.c. version of Fig. 6(b). The relations between the two are easily obtained but, as the procedure is exactly the same as for the d.c. circuit and the relations so derived are the

same, it is unnecessary to repeat it.

The form of equivalent circuit shown in Fig. 9 is actually the one most used in the literature, although it is usually derived for the earthed-base circuit rather than the earthed-emitter. Generally, the one of Fig. 8 is to be preferred because it leads to simpler design equations. The one of Fig. 9, however, has the merit that it is an obvious equivalent circuit of a thermionic valve plus a pair of resistances. In other words, a real valve circuit can be built which will simulate an earthed-emitter transistor.

In Fig. 9, the elements ρ_c and $i_b \rho_m$ are the same as the equivalent circuit of a valve if $\rho_c=r_a$ and $i_b\rho_m=\mu v_g$. If they are replaced by a valve, ρ_e becomes a cathode resistance and ρ_b a grid-cathode resistance

across which v_g is developed by i_b , so $v_g = i_b \rho_b$, and therefore, $\mu = \rho_m/\rho_b$. The circuit thus has the form shown in Fig. 10.

This circuit is a good one as an analogy for an n-p-n transistor for its characteristics are obvious to anyone versed in valve circuits whereas those of a

transistor are not so readily apparent.

It is instructive to insert some numerical values. A junction transistor may have $\rho_b = 750~\Omega$, $\rho_c = 45~\mathrm{k}~\Omega$, $\rho_e = 35~\Omega$ and $\rho_m = 1.5~\mathrm{M}~\Omega$. The equivalent value lent valve may thus have a resistance of 45 k Ω which is quite feasible and a μ of 1,500,000/750 = 2,000, which is rather impracticable, for it means a mutual conductance of 2,000/45 = 44.5 mA/V.

The transistor is thus equivalent to a superlatively good valve spoilt by a very low input resistance. It is so much spoilt, in fact, that the overall gain is no better than that of quite an ordinary valve.

So far, we have said little about the point-contact transistor. In fact, the same method of approach is possible and the same equivalent circuits are applicable. In practice, however, difficulties arise in obtaining characteristics of the same form as those of Fig. 2 because, in the earthed-emitter connection, the pointcontact transistor can have negative input and output resistances. It needs careful use in this circuit if

it is to be stable.

Because of this, it is usual to plot its characteristics for the earthed-base connection, in which it is in-herently stable, and, because this is done for the point-contact transistor, it is quite common to do it also for the junction types. Transistor characteristics and constants are more often published for the earthed-base connection than for the earthed-emitter, however the transistor may be used. It is necessary, therefore, to consider the earthed-base transistor in some detail and this we shall do in Part 3.

(To be continued)

L.E.E. Awards to Authors

THE major premium of the Institution of Electrical Engineers for a paper read or accepted for publication during the last session—the Institution Premium (value computing the last session—the Institution Fremium (value £50)—is to be given to Dr. D. M. MacKay, of London University, author of "High-speed electronic-analogue computing techniques." The John Hopkinson premium (£25) goes to Dr. M. J. Kelly and G. W. Gilman (Bell Telephone Labs.) and Sir Gordon Radley and R. J. Halsey (G.P.O.), authors of the paper "A transatlantic telephone cable."

Dr. N. W. Lewis (G.P.O.) is awarded the Blumlein-Browne-Willans premium (£20) for his paper "Waveform responses of television links." The Fahie premium (£10) goes to J. M. C. Dukes (S.T.C.) for "The effect of severe amplitude limitation on certain types of random signal," and the Webber premium (£10) to W. E. Willshaw (G.E.C.), Dr. H. R. L. Lamont (R.C.A.) and E. M. Hickin (G.E.C.) for "Experimental equipment and technical and the control of the co niques for a study of millimetre-wave propagation.

The premiums to be awarded for papers presented to the Radio Section are: Duddell (£20) to E. G. Rowe, P. Welch and W. W. Wright (S.T.C.) for "Thermionic valves of improved quality for Government and industrial valves of improved quality for Government and industrial purposes"; Ambrose Fleming (£10) to Dr. P. E. Axon, C. L. S. Gilford and D. E. L. Shorter (B.B.C.) for "Artificial reverberation"; a £10 premium to H. Page and G. D. Monteath (B.B.C.) for "Vertical radiation patterns of medium-wave broadcasting aerials"; a £5 premium to M. W. Gough (Marconi's) for "Some features of v.h.f. tropospheric propagation," and another £5 premium jointly to R. C. Glass, G. D. Sims and A. G. Stainsby (G.E.C.) for "Noise in cut-off magnetrons."

Transmitting Colour Information SIGNALLIN

SIGNALLING TECHNIQUE IN THE BRITISH N.T.S.C. SYSTEM

ANY people are of the opinion that when a colour television system is finally adopted for this country it will be a version of the American N.T.S.C. compatible system, scaled down to fit British standards.* The ultimate choice of system largely depends, of course, on the recommendations of the Television Advisory Committee, and they will no doubt have several alternatives to consider when the time comes. One possible candidate, for example, could be the frame-sequential system and another the version of the N.T.S.C. system in which the colour information is transmitted in a different channel (or even band) from the brightness information. Be that as it may, the T.A.C. have already voiced the opinion (some may think quite wrongly) that British colour television ought to be compatible, which is as good as saying, in the present state of the art, that it ought to be something very much like the N.T.S.C. system, with the colour transmitted either inside or outside of the monochrome band. Undoubtedly this would be popular with the B.B.C., because it would involve very little change to their existing black-and-white transmitting equipmentonly the addition of colour circuits. It may be significant, too, that the two major demonstrations

of colour television so far, by the rival firms Marconi and E.M.I., have both been using compatible systems.

On the assumption that a version of the N.T.S.C. system may well be adopted in Britain (and Wireless World is not necessarily in sympathy with the idea), it would perhaps be worth while looking at some of the aspects of the system which have not been fully explained in this journal so far. These are mostly to do with the processing of the colour information for transmission—a business which is quite significant in that it has some bearing on the design of domestic colour receivers. The transmission of the brightness information is not so important because it is virtually the same as the transmission of blackand-white pictures in our existing system-and in fact the brightness channel of colour receivers would look very much like the whole of the present monochrome receivers.

Wireless World has already explained the basic principles of the N.T.S.C. type of system* but it may be as well to recapitulate some of the characteristic features. At the transmitting end the colour information from the colour camera is separated into two main components, a signal conveying the brightness information of the picture and a signal

^{* &}quot;American Colour Television," Wireless World, November. 1953

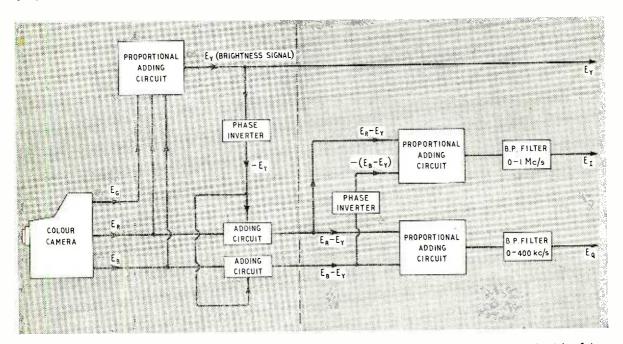


Fig. 1. Block schematic (very simplified) of the transmitting end of the British N.T.S.C. system. The section to the right of the broken line contains the circuits for producing the modified colour-difference signals from the original ones (to the left of the line)

^{*} See for example "Colour Television on 405 Lines" Wireless World. June, 1954

carrying the colour information (i.e. hue and satura-The brightness signal is received on existing monochrome television sets as a black-and-white picture, while in colour receivers it is recombined with the colour signal to produce a complete colour picture. We explained how the colour information was produced in the form of two colour-difference signals, $E_R - E_Y$ and $E_B - E_Y$ (as shown on the left of Fig. 1), which were transmitted as amplitude modulation on two components of a sub-carrier having a phase difference of 90° between them. The two components were combined, as shown by the vector diagram Fig. 2(a), to produce a single colour signal which carried hue information by phase modulation and saturation information by amplitude (Fig. 2(b) shows the various phase modulation. angles of the colour signal and the colours which they represent.) The sub-carrier was actually suppressed and the remaining sidebands were transmitted within the same band as the brightness signal by a process of frequency interleaving.

This description was actually of an early version of the N.T.S.C. system. The system which is now being used is basically the same, but the two colour-difference signals have been slightly modified. They are now each composed of certain proportions of both of the original colour-difference signals. These proportions are arranged so that one modified colour-difference signal conveys colour information ranging from orange to bluish-green (cyan) and the other colour information ranging from yellowish-green to purple. In the British version of the N.T.S.C. system* the orange-cyan signal is allowed a bandwidth of about 1 Mc/s while the yellowish-green-purple signal is restricted to about 400 kc/s.

Human Colour Vision

This modification to the system is actually a method of processing the colour information which takes advantage of the limitations of the eye's colour vision to secure the best possible transmission conditions within the limitations of the signalling channel. Our November 1953 article on the N.T.S.C. system has already explained that the human eye is insensitive to colour in very fine detail and only per-ceives it in the form of brightness changes. The N.T.S.C. system therefore does not transmit the very fine detail of the picture through the colour channel, but leaves it to the brightness channel, which has the full 3-Mc/s bandwidth available to handle it. With very coarse detail the eye is able to perceive the colours properly, but as the coarse detail is represented by a video signal of low information-content the bandwidth of the colour channel here can be quite small—in fact about 400 kc/s.

In between these two extremes there is a grade of moderately fine picture detail where the eye is only partly effective in its colour perception. It is aware of the existence of colours but cannot distinguish between them properly—in fact it is partly colour-blind. For example, blue and yellow are confused with grey, brown is difficult to distinguish from crimson, and blue is confused with green. Reddish colours, however, remain clearly distinct from blue-greenish colours, and, in fact, the eye tends to interpret all other colours in terms of these two opposites, or mixtures of them. With this moder-

* "Colour Television on 405 Lines," Wireless World, June, 1954.

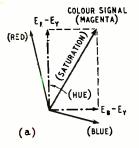
ately fine grade of picture detail, then, colour vision degenerates from being a three-colour process to a two-colour process.

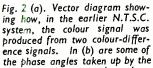
The effect can be illustrated in graphical form on the well-known Maxwell colour triangle, Fig. 3. Here the three primary colours are at the three corners, while other colours produced from mixtures of them can be specified by the spatial positions of points within the triangle. With two-colour vision, this diagram becomes nothing more than a straight line, as shown, running from orange to cyan across the original triangle. All possible colours in this two-colour system can then be specified by distances along the line instead of by spatial positions within the two-dimensional diagram. (It is interesting to note that orange and cyan are the two basic colours which have been found by experience to give the best approximation to reality in two-colour photography.)

The orange-cyan line, then, represents the "working characteristic" of the eye on colour detail of medium fineness. In the British N.T.S.C. system this particular grade of detail is represented by a band of video frequencies of up to about 1 Mc/s. The transmission circuitry is therefore arranged so that colours in the orange-cyan range are conveyed by a particular signal—the modified colour-difference signal, as mentioned above-which is, in fact, allowed this 1-Mc/s bandwidth. The other modified colourdifference signal conveys the remaining range of colours, yellowish-green to purple (. lso shown in Fig. 3) which, when combined with the orangecyan range, give complete three-colour reproduction. But as the eye is only sensitive to three-colour reproduction in very coarse detail it is possible to transmit this yellowish-green to purple signal with quite a narrow bandwidth, in fact about 400 kc/s. Thus, considering the system as a whole, one can see that the transmitted picture information is divided into three categories—coarse picture detail (up to 400 kc/s) which is transmitted in full colour; moderately fine detail (up to 1 Mc/s), transmitted in two colours; and very fine detail (up to 3 Mc/s) which is only transmitted as brightness changes as in an ordinary monochrome system.

The actual method of producing the modified colour-difference signals can be seen from Fig. 1. In the first place, the brightness signal, E_{γ} , which provides the black-and-white picture for existing receivers, is formed by adding together certain proportions of all three primary-colour-component signals from the camera. The colour difference signals E_R-E_{γ} and E_B-E_{γ} are obtained, as explained in the November 1953 article, by adding a phase-inverted version of E_{γ} (that is $-E_{\gamma}$) to E_R and E_B , the purpose being to remove the redundant brightness information which exists in E_R and E_B . Green is not transmitted separately because it can be obtained at the receiver by subtracting the sum of the red and blue signals from the brightness (or "white") signal.

More adding circuits are now brought into play to form the modified colour-difference signals from the original ones. First of all the orange-cyan signal. To produce this, 74% of $E_R - E_r$ is added to 27% of a negative quantity, $-(E_B - E_r)$, which is obtained by a phase inverter from the positive $E_B - E_r$ signal. This $-(E_B - E_r)$ signal in fact represents the opposite or complementary colour of blue, which is yellow, and one can see that 27% of yellow added to 74% of red





colour signal and the colours they represent. (Red is displaced from $E_R - E_Y$ because it is a resultant of $E_R - E_Y$ and $E_B - E_Y$ and although E_B is zero the $-E_Y$ makes $E_B - E_Y$ into a minus quantity.)

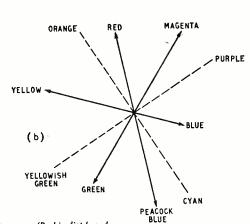
will produce orange. Thus the modified colour-difference signal coming out of the adding circuit now represents the orange in the picture (or colours seen as orange) directly, whereas before it was conveyed indirectly by two separate components in the $E_n - E_n$ and $E_n - E_n$ primary colour channels.

 $E_R - E_Y$ and $E_B - E_Y$ primary colour channels. It is easy to see that yellow and red added together will produce orange, but what about the cyan end of the "working characteristic" in Fig. 3? When colours in this part of the spectrum (or colours which the eye interprets as cyan) appear in the picture $E_R - E_Y$ becomes a negative quantity $-(E_R - E_Y)$, which represents the opposite of red, or peacock blue, while the $-(E_B - E_Y)$ goes back to positive, which of course represents blue. The added proportions of the two then produce cyan, which lies between blue and peacock blue on the colour triangle. In other words, the colour signal vector which represented orange in the first instance, at "11 o'clock" in Fig. 2(b) undergoes a 180° change of phase and appears at "5 o'clock" between blue and peacock blue. For the other modified colour-difference signal (yellowish-green to purple) the two constituents are 41% of $E_B - E_Y$ plus 48% of $E_R - E_Y$. When these values are positive they add to give the purple ("2 o'clock" on Fig. 2 (b)) and when they are negative they give the yellowish-green ("8 o'clock" on Fig. 2 (b)).

After the two signals are formed they are passed through band-pass filters as shown. The result is that both of them are effective and give three-colour reproduction for video frequencies up to 400 kc/s (coarse picture detail), while at video frequencies between 400 kc/s and 1 Mc/s (moderately fine detail) only the orange-cyan signal is transmitted, giving the two-colour reproduction to which the eye is physiologically restricted in this range. As was mentioned above, the two colour-difference signals in the earlier N.T.S.C. system were transmitted by being modulated on to two components of a subcarrier, displaced 90° in phase, which were then combined to form a single r.f. signal (Fig. 2 (a)). The modified colour-difference signals in the British system are handled in exactly the same way, the orange-cyan signal being designated E_I and the yellowish-

green to purple signal E_Q .

At this point one might be inclined to ask: why bother to restrict the E_Q signal to about 400 kc/s bandwidth when the E_I signal is already causing the



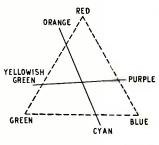


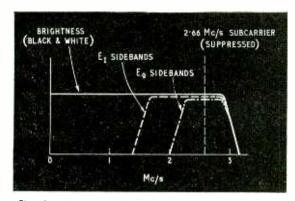
Fig. 3. Maxwell colour triangle with two axes showing the ranges of colours transmitted by the two modified colour-difference signals E₁ and E₀.

final r.f. colour signal to take up a full 1 Mc/s? The answer to this is bound up with the problem of avoiding mutual interference and cross-talk between the various component signals in the complete transmission channel (see Fig. 4).

First of all it is essential that the colour signal, which is transmitted within the same 3-Mc/s band as the brightness signal, shall have minimum visibility on the screens of black-and-white receivers. One of the expedients necessary to achieve this is to make the colour sub-carrier frequency as high as possible so that the pattern it produces on the screen will have a very fine structure. Now with a high subcarrier frequency, placed well to the right as in Fig. 4, it is clearly only possible to transmit the lower sidebands of a 1-Mc/s colour signal in full, the upper sidebands being partly removed by the upper limit of the 3-Mc/s pass band. If both E₁ and E₂ were given the full 1-Mc/s bandwidth they would both have sidebands like E₁ in Fig. 4. Transmission of these two signals on the same sub-carrier would be satisfactory over the small band where double-sideband operation is possible, but beyond this point the missing upper sidebands would have the effect of introducing spurious signals into the E, channel from the Eo channel and vice versa. The result on the receiver screen would then be incorrect colour reproduction at the edges of objects.

Overcoming Colour Cross-Talk

By restricting the E_Q signal to a bandwidth of about 400 kc/s, however, this cross-talk problem is overcome, simply because over the range where E_I consists only of single (lower) sidebands (400 kc/s to 1 Mc/s), there is no E_Q signal for it to interfere with. In other words, the two colour-difference signals E_I and E_Q are transmitted together on the same sub-carrier only in the video frequency range where double-sideband operation is possible for both, and consequently no interference occurs between The sub-carrier frequency is positioned so that this d.s.b. range is big enough to accommodate the band of video frequencies for which three-colour reproduction is effective. Beyond 400 kc/s a single colour-difference signal is transmitted, E_I, in the form of a set of lower sidebands, and only this one signal is necessary because the viewer's eye cannot perceive anything more than the two-colour information which it conveys. In the British N.T.S.C. system the sub-carrier is placed at 2.66 Mc/s, and



Frequency characteristic of the 3-Mc/s vision channel of the British N.T.S.C. system, showing how the sub-carrier and its sidebands carrying the colour information are placed relative to the upper limit of the pass band.

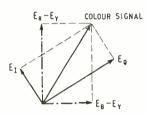


Fig. 5. Vector diagram of the $E_{\rm I}$ and $E_{\rm Q}$ signals and their resultant colour signal, showing how the colour signal vector can also be separated into components equivalent to $E_R - E_Y$ and $E_B - E_Y$. The angles taken up by the colour signal vector with the various colours are the same as in Fig. 2 (b).

this represents a compromise between two conflicting requirements—mitigating the interference pattern on the screens of black-and-white receivers and avoiding colour cross-talk.

Colour information in the British N.T.S.C. system, then (as in the American system), is transmitted not in terms of the well-known primary colours red, green and blue, but by another group of colours which are more suited to the characteristics of the signalling channel (including the human eye). In the earlier system the red, green and blue primarycolour information was converted into two colourdifference signals which were finally combined into a complete colour signal; this represented the hue of the colour by its phase angle and the saturation by its amplitude. The modified primaries in the later system are also transmitted as colour-difference signals in a similar fashion and the transmitted colour signal has the same phase angle and amplitude as before for a given hue and saturation—that is, for video frequencies up to about 400 kc/s. Between 400 kc/s and 1 Mc/s only one of the modified colour-difference signals is in operation, giving two-colour reproduction along the orange-cyan line in Figs. 2(b) and 3. The vector representing the transmitted colour signal then does not rotate through all the positions shown in Fig. 2(b), but rises and falls along the line marked ORANGE-CYAN (in both positive and negative directions).

As far as the receiver is concerned, then, it is presented with a colour signal which is modulated in both phase and amplitude, and from these modulations the original colour information has to be recovered. The technique usually adopted makes use of synchronous detectors. The incoming sidebands are mixed with two local oscillations which have the same frequency as the colour sub-carrier but are displaced 90° in phase (as at the transmitting end). One mixer then recovers the E, signal as a product while the other recovers the E_Q signal, and from these the original E_R , E_B and E_G primary-colour-component signals are eventually obtained.

It is important, however, that the two local oscillations shall have the same phases as the two sub-carrier components at the transmitting end, otherwise they will not recover the E_I and E_Q signals but something else. The point is illustrated by the vector diagram Fig. 5, which shows the final colour signal and the E₁ and E_Q signals from which it is composed. Here, as in Fig. 2(a), the colour signal vector is a resultant formed by the two component vectors E_I and E_Q , and these component vectors indicate the phases which the oscillators at the receiving end must have if E_I and E_O are to be recovered. It is clear, however, that the resultant vector can be separated into many other pairs of components, apart from E_I and E_Q , all having the 90° phase displacement between them. In practice this can be done at the receiver by altering the phases of the two synchronous-detector oscillators -and, in fact, these phases can be shifted so that the detectors recover not E_I and E_Q but two components equivalent to $E_R - E_Y$ and $E_B - E_Y$ in Fig. 1, the original

colour-difference signals.

Thus, at the receiving end, the colour information can be obtained in two different forms, either as E, and E_Q signals giving wide-band colour information, or as $E_R - E_Y$ and $E_B - E_Y$ signals giving narrow-band colour information. The advantage to be gained by recovering the narrow-band signals is simplicity of receiver design and hence cheapness in manufacture, and most of the present American receivers are, in fact, using this system. Moreover, R.C.A. in America have recently produced an improved colour demodulator for their receivers which not only recovers the $E_R - E_Y$ and $E_B - E_Y$ signals but the $E_G - E_Y$ signal as well, and also gives enough output to drive the tri-colour c.r. tube directly. It uses only one double valve and is undoubtedly a great advance in circuit simplification. At the same time, of course, this type of operation will not give the high-definition colour reproduction that can be obtained by using more elaborate circuitry to recover the E_I and E_Q signals. However, if the colour information is transmitted in the way described above it will be possible to use the cheaper receivers (a real necessity for this country) and at the same time leave the door open for more complex receivers giving better colour reproduction if they are ever wanted.

Broadcast Receiver Sales

AMENDED figures for the sale of broadcast receivers in April (see p. 346, July) have now been issued by the British Radio Equipment Manufacturers' Association, together with the figures for May. It will be seen from the table that the retail sales of television receivers in May dropped by 11,000 (15%) compared with the previous month and by nearly 40% compared with the January figure.

		Sound	Radiograms	Television
January February March April May	 	98,000 99,000 95,000 79,000 73,000	35,000 33,000 24,000 16,000 15,000	103,000 98,000 85,000 75,000 64,000

-And Its Advantages for Band-III Reception

HE present seems an appropriate time to say something about the cascode, because although it is not at all new it has only just begun to be sold to the public. In fact, the said public, as a class, are still blissfully unaware of the infiltration of cascodes into the privacy of the home. All they know is that they are foresightedly taking steps to re-establish the ascendency over the Jones's that they lost when the Jones's too installed television. These steps consist either in buying a new model fitted for Band III, or a box of tricks to adapt their present model; and they hope that, by the time the Jones's realize that advertisement TV has actually begun, the waiting list for equipment to enable it to be seen will be very long. But foresighted though they may be, unless they are technically minded they still won't realize that they have bought a cascode.

Wireless World readers, however, may be more interested in the why and how of the cascode than in knowing which washing powder makes that dainty

woolly newer than new.

The name "cascode" dates from before the war,* and may perhaps be regarded as an abbreviated form of "casc(aded-triode amplifier having characteristics similar to, but less noisy than, a single pent)ode." But that original arrangement was slightly different from the present form; and was devised for an altogether different purpose, at the extreme opposite end of the frequency scale—zero and thereabouts. The cascode has also been highly recommended for audio frequencies.† So altogether it is a versatile creature, and we ought to know something about it.

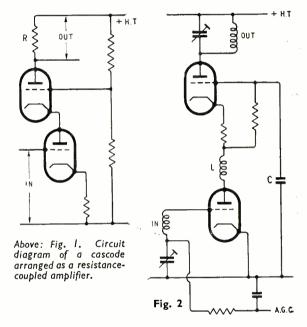
Fig. 1 shows a cascode arranged as a resistance-coupled low-frequency voltage amplifier. Essentially the same circuit could be used as the amplifier in a voltage-stabilized power unit. Fig. 2 shows a cascode arranged as a v.h.f. amplifier in a Band-III television receiver or adapter. There are all sorts of variations on the same theme, so the first thing we want to settle is the basic theme itself—the minimum that constitutes a cascode.

The short answer is: two stages of amplification, comprising an earthed-cathode triode, cascade-connected to an earthed-grid triode. But perhaps that answer itself needs a little amplification.

First of all, "cascade-connected." That means connected one after the other, as distinct from valves in push-pull and parallel stages, which come two at a time. It is not—in spite of the appearance of Fig. 1—quite the same thing as series connection. In cascade connection, which is the usual way of connecting successive valves in an amplifier, the output of the first provides the input of the second.

Now there are three basic ways of connecting a

single valve to make a stage of amplification. These ways can conveniently be named according to the electrode that is earthed; the other two electrodes are those used for input and output respectively. The commonest arrangement of the three is the earthed-cathode, in which the input is led to the grid and the output taken from the anode. The next commonest is the earthed-anode, better known as the cathode follower. It, too, has the input connected to the grid, but the output from the cathode. Lastly, the earthed-grid (called by crypto-Americans



Above right: Fig. 2. Circuit diagram of a cascode arranged as a v.h.f. amplifier, say for Band III television.

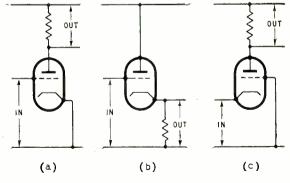


Fig. 3. Essentials of the three ways of connecting a triode as a stage of amplification: (a) earthed cathode; (b) earthed anode; (c) earthed grid.

Wireless World, August 1955

^{* &}quot;Electronic Voltage Stabilisers," by Hunt and Hickman, Rev. Sci. Inst. Jan. 1939, p. 6.

^{+ &}quot;A Low-noise Amplifier," by Wallman, Macnee and Godsden Proc. I.R.E., June 1948, p. 700.

^{‡ &}quot;Cascode Audio Amplifier has Low Noise Level, " by R. L. Price, Electronics, March 1954, p. 156.

"grounded-grid"), in which the grid (however did you guess?) is earthed, the input goes to the cathode, and the output comes from the anode. Fig. 3 lines up these three, stripped of all practical details, for inspection.

In a two-valve cascade-connected amplifier, the first stage can be connected in any of these three ways. Each of these three varieties can be subdivided into three, according to the way the second stage is connected. So altogether there are nine possible combinations. When this fact first dawned on my consciousness, I supposed that most of the combinations would be of purely academic interest, only two of three being used in practice; but on going steadily through the lot I found that most, if not all, had been used at some time, for some purpose. For instance, did you know that there was actually a special type of valve—the 6B5—embodying the earthed-anode earthed-cathode combination? Then the earthedanode earthed-grid combination is the basis of the well-known coupled-cathode or "long-tailed pair," used in many valve voltmeters and oscilloscopes. The earthed-grid earthed-cathode is used in radar and other v.h.f. receivers. But however interesting it might be to explore all nine, the only one on to-day's schedule is, as already mentioned, the earthed-cathode earthed-grid combination.

Coupling Impedance

You will of course want to know why anybody should choose this in preference to the earthed-cathode earthed-cathode, which gives easily the most amplification of any of the nine. But before going into that, let us just make sure that Figs. 1 and 2 really are earthed-cathode earthed-grid cascade combinations and not just the valves connected in series that they appear to be. Looking again at Fig. 3 we see that all three circuits include a resistor (or some other kind of impedance) which performs two roles: providing a path for the steady feed current through the valve, and by its impedance causing the signal current to set up a signal output voltage. This output voltage can then be applied to the input of a following stage. If this following stage is of either the (a) or (b) types, the coupling impedance is still required for the first stage; but type (c) is unique, because not only do its input terminals provide an impedance across which the input voltage can be set up, but the cathode terminal in particular also provides a source of feed current for the first stage. So the coupling impedance can be omitted.

As drawn in Fig. 1, the second or upper valve appears most clearly in its role of coupling impedance and steady-current feeder, corresponding to the resistor in Fig. 3(a). To emphasize its other role of second stage in a two-valve amplifier, I have redrawn Fig. 1 as in Fig. 4. It is now clearly the same as in Fig. 3(c), the only apparent difference being the point to which the grid is connected; and that is the same in principle, being equivalent to earth even though, in order to bias the grid suitably relative to its cathode, it is taken to a source of positive potential. Fig. 4, although its circuit is identical with Fig. 1's, shows more clearly that, while the second valve is in series with the first for current-feed purposes, it is truly in cascade as regards signal amplification.

And now we come to the question of why this arrangement should ever be preferred to the more highly amplifying all-earthed-cathode system. The

answer depends on which of the two main applications we have in mind. The chances just now are heavily in favour of Band III being in mind, rather than voltage stabilizers, though we'll come to them too in due course.

At the very high frequencies of Band III it is difficult to generate such massive power for transmitting as on the lower frequencies. The waves, when transmitted, are more rapidly attenuated; and the receiving aerials are necessarily short and therefore limited in collecting ability. So the signal voltage, or rather microvoltage, that can be brought to the input of a receiver is likely to be very small. At the same time it must, for television, be spread over a wide frequency band and so has to compete with a lot of noise of the kind self-generated in circuits and valves. In fact, even though v.h.f. isn't too easy to amplify, the limit is not set by that difficulty so much as by the signal-to-noise ratio that can be achieved. If the signal doesn't succeed at the outset in poking its head high enough above this noise, no amount of subsequent amplification will help it, because the noise is amplified too. So all depends on the first stage of amplification, in which the signal is at its weakest.

Because of the greater difficulties of amplifying v.h.f. signals, there is an obvious inducement to change to a lower frequency as soon as possible. But a frequency-changer stage is at best a comparatively noisy affair, so doesn't make at all a suitable first stage for signals that are already barely strong enough to stand clear of noise. A preliminary stage of amplification, even if it doesn't give very much, and even if it introduces some noise itself, helps the signal to master the greater frequency-changer noise. In case this point is not entirely obvious, let us suppose that a certain frequency changer introduces $6\mu V$ of noise and neither amplification nor loss, and that a stage of amplification introduces five-fold voltage gain and the equivalent of $3\mu V$ of noise at its input. If the incoming signal is $16\mu V$, together with $2\mu V$ noise, the incoming signal/noise voltage ratio is 8:1 and its power ratio 64:1. After the frequency changer alone the signal would still be $16\mu V$ and the noise voltage $\sqrt{(2^2+6^2)}=6.3^*$, so the signal/noise power ratio would be degraded to 6.5:1. But if the amplifier were used first, its output would include $5 \times 16 = 80 \mu V$ of signal and $5 \sqrt{(2^2 +$ 3^2) = $18\mu V$ of noise (nearly 20:1 power ratio), and the output of the frequency changer following would be $80\mu\text{V}$ signal and $\sqrt{(18^2 + 6^2)} = 19\mu\text{V}$ noise, making the ratio 18:1. This simplified example shows how even a moderate amplification, itself not noiseless, can much improve the net result.

What Valve to Use

Having decided to use at least one stage of amplification at the original frequency, we are then faced with conflicting claims in the choice of amplifier. A r.f. pentode has internal screening which helps to prevent feedback via stray capacitance turning the amplifier into an oscillator, as would inevitably happen with an earthed-cathode triode owing to its anode-togrid capacitance. On the other hand, a pentode is a good deal noisier than a triode. I went into detail about this noise business in the May and June 1952 issues, so for present purposes it should be enough

^{*} Noise voltages, like a.c. voltages of unequal frequency, have to be added together in this way; see " Total Power " in the March 1952 issue.

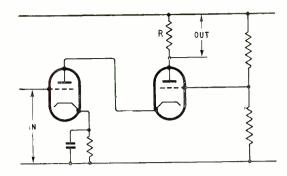


Fig. 4. Rearrangement of Fig. I to show more clearly the role of the second valve as an earthed-grid stage.

to say that because the current through a valve consists of separate electrons there is bound to be a small random fluctuation around even the steadiest average, and it is this fluctuation that is referred to as noise. In a pentode the fluctuation is greater than in a triode, because on top of this chance of slightly fewer or more electrons arriving during any instant, there is the chance of whether any particular electron will be intercepted by the second grid or go right on to the anode. Though you might not think so, this extra chanciness adds very considerably to the noise. The average for 10 different kinds of pentodes used for amplification showed 75% greater noise voltage than the average for a number of triodes. So there is a strong inducement to use a triode, if it can be made to amplify at v.h.f.

As we know, it cannot if it is simply connected in the usual earthed-cathode manner. This is where the earthed-grid triode came in, especially during the last world war for 200-Mc/s radar. Unfortunately an earthed-grid triode presents a very low input impedance, even by v.h.f. standards. As we see from Fig. 3(c), it is not much more than the *output* impedance of a cathode follower, and we know that that is very low—approximately $1/g_m$, or say 200Ω if g_m , the "slope" of the valve, is 5mA/V. The only difference between this circuit and a cathode follower is the external anode resistance, which reduces the overall slope by adding to the internal anode resistance, \mathbf{r}_a . In a v.h.f. amplifier, the external resistance is the dynamic resistance of a tuned circuit, and is likely to be considerably smaller than r_a . So the input resistance is still very low, and establishing a signal voltage across it costs plenty of signal power. Stage gain is therefore even less than usual for a triode.

The attraction of the cascode is that it combines the low noise of a triode with earthed-cathode input resistance and the high gain and stability of a pentode. And although the circuit diagrams here make it look as if two valves were being used in place of one, in practice the two triodes take the form of one double triode costing about the same as one pentode and occupying the same space and valve-holderage.

The second half of the cascode is an earthed-grid stage as just described, and its low input resistance constitutes the anode load or coupling resistance (Ra) of the first half. So although this first half is an earthed-cathode stage, it is prevented from bursting into oscillation by the fact that its R_α is too low to give enough amplification. When the R_α of a voltage amplifier is much smaller than the ra of the valve, the voltage amplification (A) is approximately equal

to $R_{\alpha}g_{m}$. But in this case we know that R_{α} is approximately equal to $1/g'_{m}$, where g'_{m} is the slope of the second triode, as modified by *its* anode load. So the amplification of the first stage is

$$A \simeq R_a g_m \simeq \frac{g_m}{g'_m}$$

 $A\simeq R_\alpha g_m \simeq \frac{g_m}{g_m'}$ The slopes of the two valves in a double triode are (or should be) more or less equal. So even if the load resistance of the second were as much as equal to r_a , g'_m would be half g_m , and A consequently no more than 2. In practice, at v.h.f., it would probably be nearer 1. This doesn't mean that the first stage contributes next to nothing; its real job is to drive the second stage without damping down the input circuit or introducing a lot of noise. It functions, in fact, very like a cathode follower.

The output of the second stage is quite conventional. Fig. 2 shows the whole cascode as arranged for Band-III amplification. Instead of being taken to a tapping on a potential divider, the grid of stage 2 has conventional bias arrangements, but at signal frequency is tied down to earth by C. This modification of Fig. 1 enables automatic gain control to be applied in the customary manner to stage 1. In Fig. 1, a.g.c. bias would be almost completely ineffective in changing the anode current and with it the amplification, because the fixed potential of grid 2 would tend to hold it constant. The only other thing that might perhaps excite curiosity about Fig. 2 is L. Its purpose is to neutralize stray capacitance at the top frequency of Band III, where there would otherwise be a falling off in amplification.

Mathematical Approach

If, as I hope, you like to work things out mathematically for yourself, I can recommend analysing Fig. 1 by writing down the equations—five of them expressing overall voltage amplification, A; signal anode current, i_a ; signal voltages of the two anodes, v_{a1} and v_{a2} ; and signal voltage between second grid and cathode, v_{g2} ; in terms of the μ and r_a of the two valves, the signal voltage v_{g1} applied to the first grid, and R. If the valves are not assumed to be identical, the answer should come out to

$$A = \frac{(\mu_2 + 1) \mu_1 R}{R + r_{a2} + (\mu_2 + 1) r_{a1}} \dots \dots (1)$$
If the valves zre identical this simplifies to
$$A = \frac{(\mu + 1) \mu R}{R + (\mu + 2) r}$$

$$A = \frac{(\mu + 1) \mu R}{R + (\mu + 2) r_a}$$
To if we is large enough

while if μ is large enough for the difference between it and $\mu + 2$ to be neglected,

$$A \simeq rac{\mu R}{R/\mu + r_a}$$
 which compares with $A = rac{\mu R}{R + r_a}$

in a conventional single-triode amplifier. These results prove what we have already gathered, that the v.h.f. amplification of the cascode as a whole is not noticeably greater than would be given by one of its valves connected as an ordinary stage if there were no such thing as Miller effect to upset its working. But there is, and the addition of the second triode overcomes it without having to fall back on the noisy pentode.

Looking at things from different viewpoints is usually a help in understanding them; and having

considered the cascode as two successive conventional stages in an amplifier, we may now care to think of it as a single unconventional stage. The difference between pentode and triode characteristics can be expressed in one way by saying that the pentode has a very much larger internal resistance ra, which is the same thing as saying that its anode current is very little affected by its anode voltage. Putting a load resistance in series with the anode therefore has hardly any effect on the amount of anode current. This is not so with a triode, however. When its anode current is increased by making the grid less negative, any anode load resistance causes the anode voltage to drop, and this cuts down the increase that would otherwise have taken place.

With this in mind, consider Fig. 1 again. The first (lower) valve has an anode load resistance consisting of the other valve plus R. If the grid of this other valve had a fixed bias relative to its cathode, the action would be as just described. But because the grid is tied to a fixed potential, the drop in cathode potential caused by any increase in anode current is equivalent to making the grid less negative, which operates to maintain the increase in anode current. Since the cathode potential cannot alter much without drastically altering the anode current in this way, it tends to stay nearly constant. This is the same

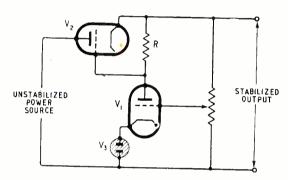


Fig. 5. Essentials of the usual type of voltage stabilizer.

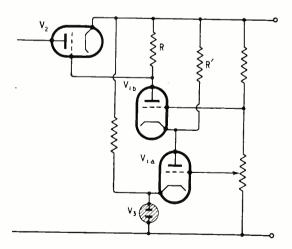


Fig. 6. Modified cascode substituted for V_1 in Fig. 5. The modification consists of R', which raises the amplification of the stage by increasing the anode current and consequently the g_m of V_{1n} .

thing as the lower valve's anode voltage staying nearly constant, regardless of changes in anode current induced by changes in its grid voltage. So the system as a whole behaves something like a pentode, in two ways: the anode current has been made to depend almost exclusively on the voltage of the first grid regardless of the presence or absence of R; and the potential of the electrode next to that grid—the first anode—has been made to stay nearly constant, and so to simulate the screen grid in a pentode, which keeps the amplifier stable.

Voltage Stabilizer Application

There is no reason, of course, why the cascode cannot be used at lower radio frequencies than v.h.f., or even audio frequencies, but amplification of lower r.f. is not usually limited by valve noise, and a pentode gives more amplification. For a.f. it is an interesting question whether, on balance, the cascode is better than the two valves of a double triode connected in the ordinary way as two earthed-cathode stages. Coming at last to the amplifier used in the usual type of voltage stabilizer, there is much more to be said for the cascode. Most of it has been said very well by V. H. Attree,* who has devised a modification that looks like establishing the cascode as undisputed king of this situation.

Just to make sure that we are both thinking about the same "usual type of voltage stabilizer," Fig. 5 is an outline diagram, for identification, in which V₁ is the amplifier in question. The greater the amplification of this stage, the more effective the stabilizer. So the choice of valve is almost invariably a pentode. To extract a reasonable proportion of the valve's potential amplification, the resistance R must be large. But the voltage across it, being the bias for V2, is normally quite small, which means that the valve is working under conditions (to wit, low g_m) that throttle most of its amplification. If two stages were used, the output voltage would be of the wrong polarity and the stabilizer would become an unstabilizer. Two stages, that is to say, connected in the usual earthed-cathode manner. But this difficulty does not apply to the cascode, because its second stage, being earthed-grid, does not reverse the polarity. The criticisms that have just been made about the pentode would go for the cascode too, if it were as per Fig. 1. But Mr. Attree has pointed out that when μ_1 and μ_2 in our equation (1) are large (because a high-µ double-triode has been chosen), and R and r_{a2} are neglected in comparison with μr_{a1} , the thing reduces to

$$\mathbf{A} \simeq \frac{\mu_1 \mathbf{R}}{\mathbf{r}_{a1}} = \mathbf{g}_{m1} \mathbf{R}$$

So the amplification depends hardly at all on the g_m of the second triode and almost entirely on that of the first, and it doesn't matter if the current through R and the second valve is small, so long as it is large enough for a good big g_m in the first. The modification therefore consists of an extra resistor to pass more current through this valve.

The circuit (less all frills) is as Fig. 6. With this improvement, the voltage amplification of the cascode can easily be well over 1,000, which makes the stabilizer incorporating it a very good stabilizer.

^{* &}quot;A Cascode Amplifier Degenerative Stabilizer," Electronic Engineering, April 1955, p. 174.

ELECTRONIC TELEPRINTING

A NEW electron-image tube that can translate coded signals from teleprinter tape or other sources into clearly defined letters and figures at speeds up to 100,000 words per minute for high-speed photographic recording has recently been developed by R.C.A. In operation the tube simulates the process of typesetting by selecting letters and figures one by one from a "font" and placing them in luminous form on the 5-in circular tube face, either in lines or in any pattern desired. The "font" is actually a lantern slide external to the tube, bearing a chart of letters and figures. An image of this is projected on to a photo-emissive cathode at one end of the tube which emits a stream of electrons in a corresponding pattern. The electron stream is then accelerated forward in the tube by a potential of 100V applied to the conductive wall coating and focused by an external coil providing an axial magnetic field.

The selection of letters and figures in the required order is accomplished by a small aperture of 0.04-in diameter at the neck of the tube which permits only one character at a time to pass through. As the electron stream pattern carrying all the letters and figures moves towards this aperture, a magnetic deflection coil (mounted inside the focus coil) shifts the stream so that only the desired character passes through and travels towards the tube face. Another set of coils then focuses and deflects the character to its proper place on the phosphor screen.

As many as 4,000 characters have been produced clearly in a single pattern on the 5-in tube face. The size of the letters and figures on the screen can be enlarged, however, by the second set of coils if required.

The deflection coils for selecting and positioning the characters contain windings for both vertical and horizontal deflection, as in television. In the teleprinting application these coils are supplied with suitable steps of current in accordance with the coded information which is "read off" (by photo-electric means) from the teleprinter tape.

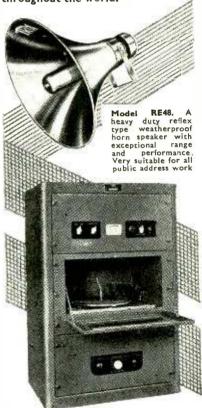
When the new tube achieves commercial form its initial applications are likely to be in electronic message transmission and in computing systems. Later it may be developed for electronic typesetting.



COLOUR TELEVISION STUDIO, claimed to be the first one built specifically for colour, and recently put into operation by the National Broadcasting Company of America at Burbank, California. An elaborate lighting system with 2,400 controls permits the pre-setting of lighting for ten scenes and also ten changes of lighting within any one scene. An unusual feature is an "audience pit" which accommodates the studio audience below floor level to avoid interference with the cameras.



There is no substitute for the unique experience of the TRIX organisation in the design and development of Sound Equipment. Each unit, whether large or small, is true to the tradition which has made TRIX workmanship respected throughout the world.



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WIRELESS WORLD, AUGUST 1955

RANDOM RADIATIONS

By "DIALLIST"

" Night Starvation "

EVEN in what passes nowadays for summertime, the mains voltage in some localities is liable to fall somewhat after, say, 8 p.m. or a little later. If the fall is not very great and the television receiver is in tip-top form, the effects of this may be hardly noticeable; but with a bigger fall or a set which has been in use for some time without renewals or replacements this "night starvation" can lead to picture shrinkage, with the typical black borders. When that happens the first thing that I suspect is the h.t. rectifier, particularly if it's of the metal type. Metal rectifiers are sound and reliable components, but they don't last for ever and it pays to renew them when tests show that their output is appreciably down. I'd estimate about two years or perhaps a bit more as their normal useful life in an average family set.

A Radio-link Problem

THE temporary East Anglian television transmitter at Tacolneston reradiates the London transmissions which it receives direct. This has been leading recently to spots of bother caused by r.f. interference from a Continental sound-broadcasting station. I haven't yet been able to find out what station it is. Under certain conditions the interference is fairly mild, merely causing a certain amount of "fish netting" on the picture. But it can at times be bad enough to blot out the picture and to make the accompanying sound almost unintelligible. To prevent viewers in the Norwich area from snowing dealers under with complaints about the misbehaviour of their perfectly guiltless receivers the B.B.C. has very wisely devised a means of letting them know when it's the transmission that is to blame: vertical white bars sent out every so often indicate to the viewer that interference is affecting the transmission.

A Monitoring Suggestion

This idea might, I think, be carried a good deal further, for TV receivers are often blamed by their users for doing things that they can't help doing. Some of the O.B. cameras, for example, "ring" quite severely, producing pronounced white outlines to the right of dark objects. And this

effect can be made worse than ever if a relay is made over a long landline. An occasional word from the announcer about this might save viewers from worrying about their sets and servicemen from having to make unnecessary journeys. I've a feeling that some, at any rate, of the monitoring should be done with ordinary domestic receivers. Those in charge would then be able to see whether any transmission was likely to cause the sets of Smith, Jones, Brown and Robinson to play up and a word of explanation (and of comfort) could be issued at suitable moments. You know the kind of thing I mean: the sync isn't always able to lock one of the scans-or maybe both of them-properly; a change of camera means a fall in the brightness level, or vice versa.

Over-Simplified?

Yes; I know that these things wouldn't happen if TV sets were a little more elaborate—and, therefore, a little more expensive. But, domestic receivers being what they are, the plain and inescapable fact is that they do happen. And that brings me to the warning given recently to the radio industry by Harold Bishop, to

whom I offer my humble felicitations well-deserved Birthday Honour. What he said in effect was that there are limits beyond which simplification and price reduction cannot reasonably be carried by television receiver manufacturers. And there couldn't be a truer word spoken. Many people feel that these things have been taken too far already. Bringing down prices is fine from one point of view but it's a far from unmixed blessing if it entails, for example, lack of d.c. restoration and d.c. amplification, plus synchronizing arrangements so poor that the picture won't lock unless the signal is bang up to the mark. Myself, I believe that large numbers of folk would gladly pay a bit more for sets that didn't suffer from these shortcomings and whose pictures remained without flutter when aeroplanes were passing by.

The Tape-Recorder Cult

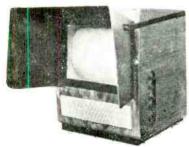
THE tape-recorder is, I suppose, the lineal descendant of the Dictaphone and other similar machines designed originally purely for office use. Today, tape recording has become a hobby whose addicts are every bit as enthusiastic as were the fans of the early days of wireless. To the musical it is of course a great joy to be able to make a record (I nearly wrote "recording"!) of any outstanding broadcast and to have it available when wanted. I must have been quite an early user of the

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Dictaphone. Years ago I used to answer a vast number of letters. It was, as you may imagine, the greatest possible convenience to be able to get a few off my chest whenever I felt so minded by dictating my replies into the machine. Once, though, there was a sad mishap. While on a fishing holiday in Devon I recorded a big batch and posted the wax cylinders to my typist. Light-hearted P.O. sorters must have had fun and games with the parcel, for it arrived with all the records in more or less powder form. That, anyhow, couldn't happen with tape.

Balanced or Unbalanced

WHEN television broadcasting was in its youth, I think I'm right in saying that the great majority of receivers were designed for use with balanced twin aerial feeders. To-day all (or very nearly all) use co-axial feeders. I'm told that this means a small reduction in manufacturing costs; if so, I wonder whether it's worth it. Interference seems to become worse and worse and a good deal can be picked up by a co-axial feeder, even though its metallic sleeve is earthed. But balanced twin, with earthed metallic screen and correct impedance matching, picks up little or none; use it and, if need be, a mains suppressor and you'll get no interference except what is actually picked up by the aerial itself. Some time ago I was using an "H" aerial over 100ft from the nearest road and near y 60ft above its surface. Motorcar interference was a nuisance with a coaxial feeder in use; a change to balanced twin, with the necessary alterations to the receiver and careful impedance matching made all the difference in the world.



TELEVISION SCREEN MASK for shielding the tube face when viewing in daylight or with the room lighting on. Made from compressed fibre material with a light-absorbent inside lining, it is fixed to the front of the set by rubber suckers and secured by an elastic strap. Available from Vendoma (TV), Station Buildings, Preston Park, Brighton, in sizes for 12-in, 14-in, 15-in and 17-in receivers.



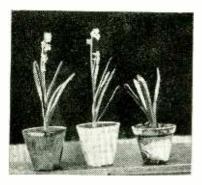
UNBIASED |

Peccavi

IN the June issue I mentioned that I was going to try to accelerate the growth of plants in my garden by exposing them to the radiations of an r.f. oscillator. I had been interested in an item of news received from the U.S.A. which said it had been noticed that in the neighbourhood of high-powered television transmitters a veritable jungle of luxuriant undergrowth had sprung up.

I found I was quite mistaken in thinking that there was anything new in this as I have been sent several photographs (one of which is reproduced) and data concerning the growth of r.f.-nurtured plants as far back as 1939. The two hyacinths on the left of the picture received radio treatment (50 Mc/s) and flowered 15 days earlier than the untreated one shown with them. I can only say I feel ashamed of my own ignorance.

In the June issue I also discussed early talking machines and a reader has tried to put me on the penitent's stool for this by implying that I said that needles and not sapphires were invariably used with disc records. To



Electronics in the garden.

prove me wrong he has sent me an Edison sapphire-using disc but adds "I see that, with true legal caution, you have left yourself a loophole."

What I certainly did not know was that some disc machines used a screwed rod to propel the sound box across the record. I thought that with discs this was invariably done by the needle running in the groove as in modern machines. This is not so, however, and the makers of the record took pride in announcing that the grooves were too shallow to have the task thrown on them of pushing the sound box along; records are made of sterner stuff nowadays.

Mummified Music

I WAS very impressed by the demonstrations given in the Royal

By FREE GRID

Festival Hall in May of the degree of hi-fi which can be achieved in mummified music by modern methods of recording and reproduction. It is given to a very few to be able to listen to a chronologically and topographically side-by-side comparison of the real thing and an embalmed version of it. Despite the fact that the hall was packed to capacity, the audience was very small compared with the number who would have liked to be there.

It is unthinkable that there should not be a repeat performance in a few months' time and I have been trying to think of a method whereby a larger audience could be reached. Mrs. Free Grid made the tom-fool suggestion that the B.B.C. should be asked to put it on the air. Apparently it did not occur to her that the musicmangling properties of the ordinary domestic wireless receiver would reduce the living and the mummified performances to the same low level; in fact, the embalmed version could be made to sound better than the real thing, for special recordings could be made to compensate to some extent for the deficiencies of the average set.

In a few years' time when there will be a hi-fi receiver in every listener's home it will be a different thing. But even to-day the number of f.m. set owners is certainly much greater than the seating capacity of the Royal Festival Hall. To hear the programme on v.h.f. would certainly not be the same as hearing it in the flesh.

All the same, I for one am willing to agree not to apply for a ticket for the next demonstration but to listen at home and let my seat be occupied by somebody less fortunate than myself who is at present beyond the reach of v.h.f. either for geographical or financial reasons. I, therefore, appeal to all other f.m. people to make a similar offer and to the organizers of the demonstration to approach the B.B.C. in the matter.

Electronic Morphimeter

I WAS interested to learn that an electronic device has been developed for gauging the depth of sleep or unconsciousness by measuring the skin resistance, which apparently varies in step with it. Surely this should have many applications other than the medical one mentioned. Apart from its obvious use in the boxing ring where it would enable the referee to see at a glance the exact condition of a recumbent pugilist it supersedes the existing form of baby alarm first described in this journal nigh on thirty years ago.

This old type alarm, as you may



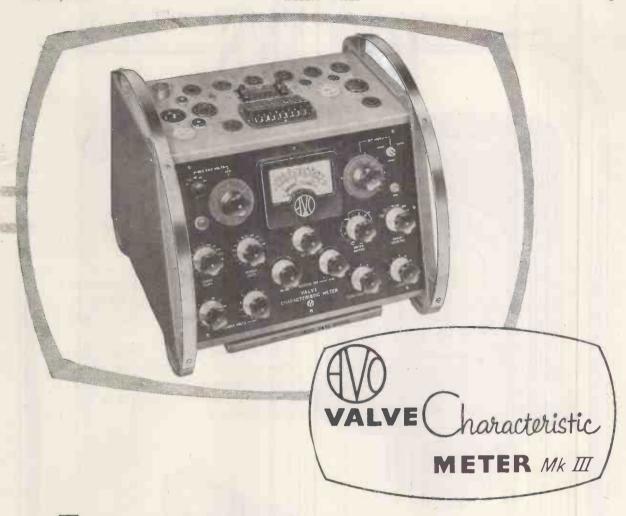
Further outlook unsettled.

remember, consisted of a mike suspended over the cot so that the nasty noises emitted by a baby are conveyed to the doting parents below. With this new device the changing resistance of the baby's skin as it approaches bawling point could obviously be caused to operate the alarm, thus obviating the mental trauma which a well-known psychiatrist has stated that babies receive when their immediate wants are not anticipated.

I would like to point out, however, that this idea of a morphimeter is not quite as new as it seems. The varying skin resistance on which its action depends is caused, I understand, by the fact that the rate of metabolism in the body changes according to the degree of unconsciousness. This changing rate of metabolism not only varies the skin resistance but has the same effect on the dielectric constant

of the body.

Long years ago I fixed up a capacitor, the two electrodes of which were the bedspring and the wire mesh of an electric blanket which Mrs. Free Grid always uses. Her body was, of course, the dielectric and I found that the capacitance of my crude device varied according to the depth of her sleep. I put this to practical use by causing it to operate an indicating meter downstairs so that on those occasions when I was detained late at the office I could avoid creeping upstairs to bed until I was assured that she was really sunk in a deep sleep.



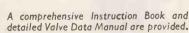
he AVO Valve Characteristic Meter Mark III offers the Radio Engineer far more than is generally implied by the words "a valve tester".

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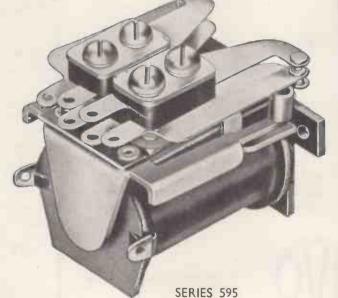
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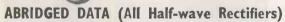
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2021	RR3—250	C¥1835	4-pin UX	2.5	5.0	{10 5	1.0	0.25 0.5	-55 to +75	
Y	RR3—1250	CV2518	B4F	5.0	7.1	10	5.0	1,25	-55 to +70	
2	RGI—240A	CV1626/1072	British 4-pin	4.0	2.7	6.5	1.25	0.25	+ 10 to + 4	
ארטס	RG3—250	CV1625	Medium Edison Screw	2.5	5.0	10	1.0	0.25	+ 10 to + 4	
2	RG3-250A	CV32	4-pin UX	2.5	5.0	10	1.0	0.25	+ 10 to + 4	
	RG3—1250	CV1629/152	Goliath Edison Screw	4.0	7.0	13	5.0	1.25	+ 10 to + 40	
MENOORI	RG4-1250	CV5	Goliath Edison Screw	4.0	11	13	5.0	1.25	+ 10 to + 4	
5	* 872A	CV642	B4F	5.0	7.5	10	5.0	1.25	+ 10 to + 4	

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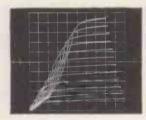


Fig. 1 Plate current - plate voltage

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Choice of four to twelve characteristic curves per family — with as many as eight positive-bias curves per family.

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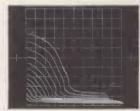


Fig. 2 Screen current - plate voltage

Series of pentode characteristic curves with grid voltage changing 2 volts/step from + 16 v to below zero, illustrating Type 570 operation with eight positive-bias curves per family. Vacuum tube is a 6AQ5, under these conditions: Plate load...300 ohms, peak plate voltage...100 v, screen-grid voltage . . . 100 v, vertical scale ... 10 ma/division, horizontal scale, fig. 1, 2, and 3 . . . 10 v/division, fig. 4, 5, and 6 . , . 2 v/division.

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Grid current - grid voltage



Fig. 5 Screen current—grid voltage

Fig. 3

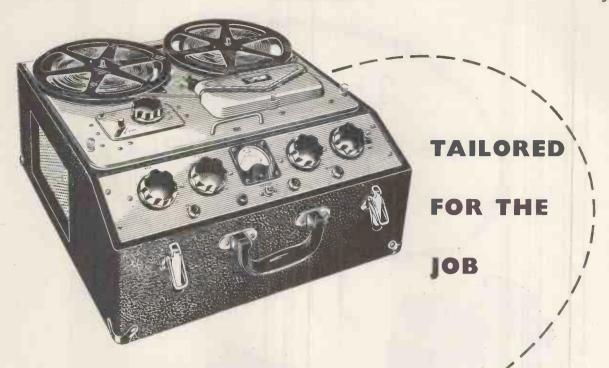
Fig. 4

Plate current-arid voltage

Grid current - plate voltage.

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It is by no means accidental that the Ferrograph has achieved so high a reputation in every country to which exportation is possible, and at a price no greater than that of an ordinary home recorder.

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Ferrograph

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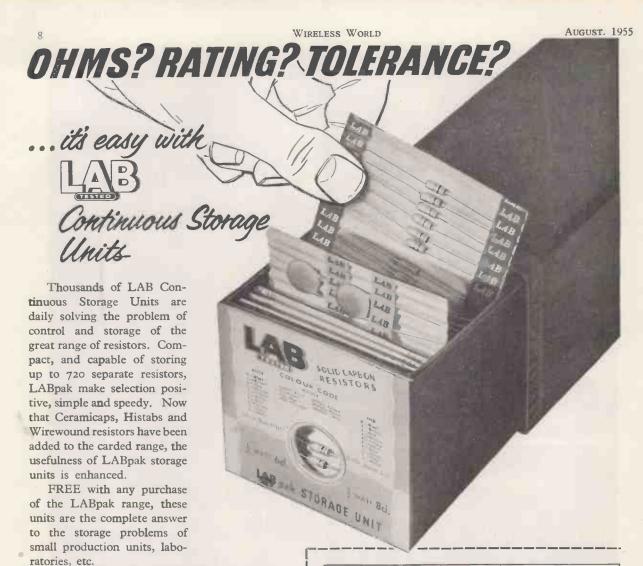


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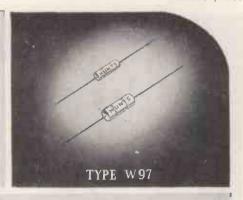
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A single metallised paper is used to wind this unit which is made possible by the use of Hunt's Patent covering the "castellated" pattern. Recent development by Hunt on a special impregnating material gives the unit remarkable brackets of operating temperature.

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The capacitors are supplied without an insulating medium on the case. If specially requested they can be supplied with an approved plastic sleeve which increases the dimensions by 0.07" in length and 0.03" in diameter. **TERMINATIONS**

TERMINATIONS
The terminations are of 24 gauge tinned phosphor bronze wire having a nominal length of 1½". Special attention is paid to the retinning of the wires after the capacitor is fully processed. Connection is made to the unit by applying copper spray to the metallising. The pigtail is soldered to this bond giving a perfect connection of exceptional strength.

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W97 "Thermetic" Midgets have a very high self resonant frequency—the following figures are quoted as a guide. 50 pF at 600 volts, which is the lowest capacitance in the range, has a self resonant frequency of 280 mega-cycles. At the other end of the range, 0.04 uF 200 volts, which is the

cycles. At the other end of the range, 0.04 µF 200 volts, which is the maximum capacitance, it is 8.5 megacycles.

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This is measured at working voltage at a temperature of 20°C. The minimum capacitance in the range, 50 pF at 600 volts, has an insulation resistance greater than 2,000,000 megohms. The maximum capacitance in the range 0.04 µF at 200 volts, has an insulation resistance greater than 25,000 megohms. The intermediate capacitances are approximately pro

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BM7	0.002	0.610	0.135
BM8	0.004	0.640	0.135
BMII	0.004	0.500	0.180
BM9	0.005	0.610	0.135
BM12	0.005	0.500	0.180
BM13	10.0	0.500	0.180
BMI4	0.02	0.610	0.180
8M15	0.03	0.610	0.260
BM16	0.04	0.610	0.260

	400 volts D.C.	Nkg.	
BM4	0.0004	0.610	0.135
BM5	0.0005	0.610	0.135
BM6	0.001	0.610	0.135
BM18	0.002	0.500	0.180
BMI9	0.003	0.500	0.180
BM20	0.005	0.610	0.180
BM21	10.0	0.610	0.260

	600 volts D.C.	Wkg.	
BM25	50 pF.	0.500	0.180
BMI	1000,0	0.610	0.135
BM26	0.0001	0.500	081.0
BM2	0.0002	0.610	0.135
8M27	0.0002	0.500	0.180
8M28	0.00022	0.500	0.190
BM29	0.00025	0.500	0.180
BM3	0.0003	0.610	0.135
BM30	0.0003	0.500	0.180
8M36	0.0004	0.500	0.180
BM31	0,0005	0.500	0.180
BM32	. 0.001	0.500	0.180
BM33	0.002	0.610	0.260
BM34	0.003	0.610	0.260
BM35	0.004	0.610	0.260



20 Mc/s FREQUENCY MONITOR

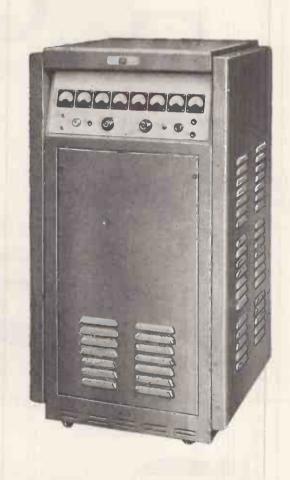
The Automatic Frequency Monitor (20 Mc/s) is but one of a series of high grade monitors now in course of manufacture for the accurate measurement of frequency.

Employing hard valve techniques throughout, it will measure any frequency in the range 10 c/s to 20 Mc/s to an accuracy within \pm 1 part in 10^6 .

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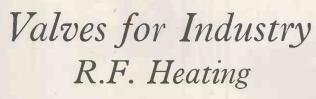
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Type	imum Operating Frequency Ill Rating (Mc/s) Cathode	Cathode	Dimensions (maximum)	Filament .	Maximum Anode Voltage kV.(D.C.)	Maximum Anode Dissipation (Kilowatts)	Useable Emission (Amperes)	Amplification Factor	Mutual Conductance (mA/V)		
BR, 1102	Maximum Frequat Full Rat	Th.	Teng mm	Diameter mm.	v. Volts	Amperes	12·0	Max O.o.	45	42	
BW.1102 BR. 1103 BW.1103	50 100 100	Th. Th. Th.	473 356 343	152 203 116	8·2 6·0 6·0	230 120 120	8·5 8·5	20.0 10.0 10.0	45 16 16	42 25 25	20·0 20·0 8·3 8·3

NOTE: CATHODE:—Th. denotes THORIATED TUNGSTEN.
BR. denotes AIR-COOLED.

BW. dengtes WATER-COOLED.

Technical data sheets are available on request.

ENGLISH ELECTRIC VALVE CO. LTD.



Waterhouse Lane, Chelmsford Telephone: Chelmsford 3491





60% brighter pictures
more contrast
extra tube life

AN Ediswan Mazda aluminized picture tube gives a picture 60% brighter and more contrasty than is possible with an ordinary tube.

In addition, Ediswan aluminizing protects the screen from ion burn and, with the new Ediswan ion trap tetrode gun to protect the cathode, tube life is increased.

Ediswan production methods, which include the special in-line vacuumizing system, ensure a higher, more uniform standard of lasting efficiency. For complete satisfaction demonstrate and recommend Ediswan Mazda aluminized picture tubes.

EDISWAN

ALUMINIZED CATHODE RAY TUBES

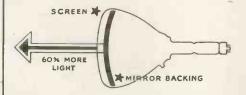
THE EDISON SWAN ELECTRIC COMPANY LIMITED, 155 Charing Cross Road, London, W.C.2 and Branches.

Member of the A.E.I. Group of Companies.



WITHOUT ALUMINIZING

Without aluminizing, tubes waste half their light (see diagram above). To counteract this the brilliance must be increased and the tube life is shortened.



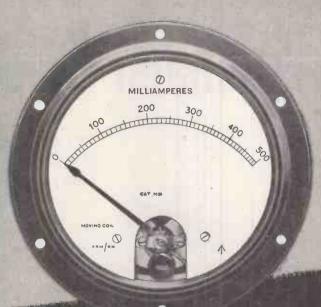
WITH EDISWAN ALUMINIZING

Ediswan aluminized tubes have a mirror backing to the screen. All the light is thus thrown forwards giving brighter, clearer pictures and extra life.

NATION WIDE SERVICE

6 fully equipped cathode ray tube service depots provide better, quicker tube testing should the need arise. Stocks of tubes are available in 26 Ediswan Offices. Only Ediswan give such complete backing to the Trade.

RVS



FERRANTI Sealed Instruments



The Ferranti 31' hermetically sealed instrument in steel case with fixing flange has received Design Approval from the Joint Service Radio Components Standardisation Committee.

This instrument is designed to comply with RCS.231 and RCL (PROV) 231

FERRANTI LTD

MOSTON

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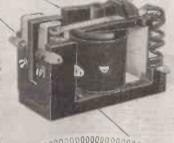
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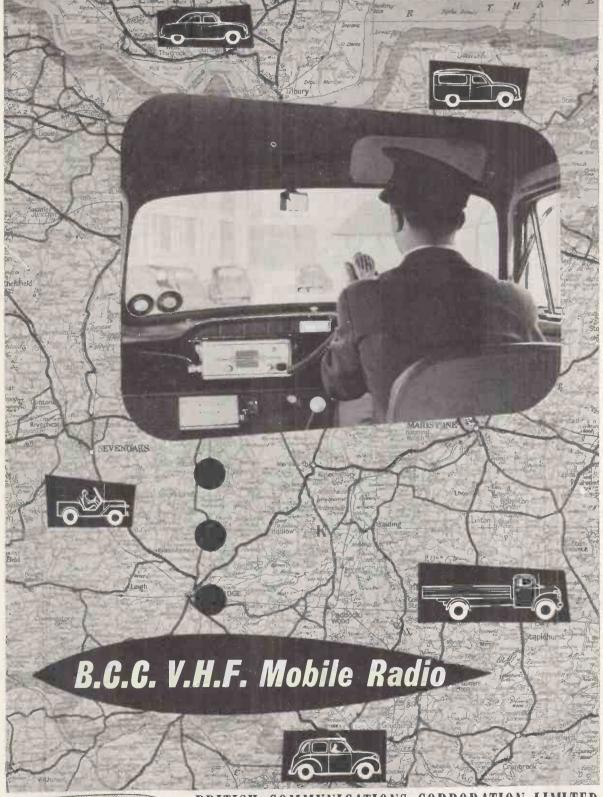
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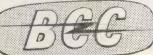
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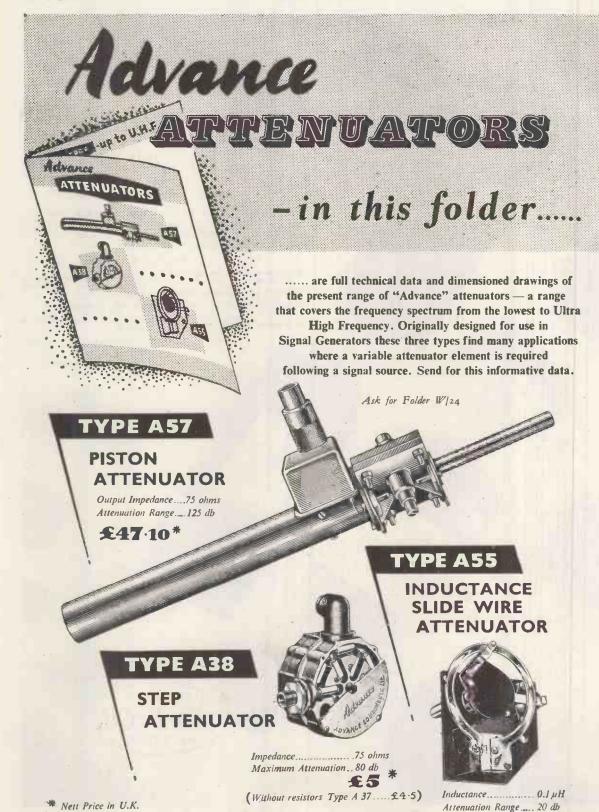
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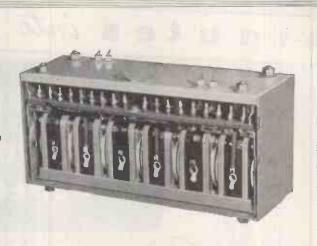
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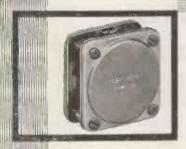
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m-i-n-u-t-e-s into seconds...

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STAR FEATURES

- ★ Heats up from cold in 6 seconds—by a light thumb pressure on the switch ring.
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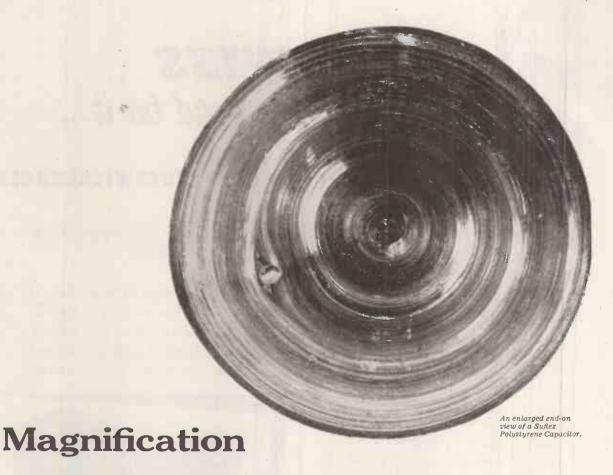
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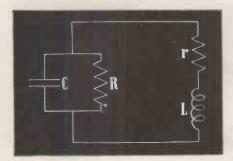
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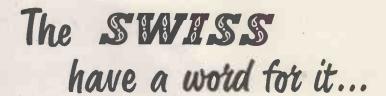
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KATHODENGEGENKOPPLUNGSWICKLUNGEN

a singularly factual one describing the design of the output circuit of the QUAD II Amplifier.

In other words the output transformer includes cathode as well as anode and output windings. Thus, the screen characteristics of the valves are used as well as those of the anode, while the voltage appearing across the cathode winding is also applied in the grid circuit as overall negative feedback without either sacrificing gain wastefully or increasing the load on the previous stage.

The resulting decrease in distortion and increase in efficiency have contributed considerably to the success of Acoustical Amplifiers over the last ten years, and are reflected in the rather unusual performance-for-size of the QUAD II.

For a full description of this circuit from its basic principles see "Amplifiers and Superlatives" Wireless World, Sept. 1952.





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No. 1 "SYMPHONY" AMPLIFIER is a 3-channel 5-watt Gram/Radio Amplifier with asconishingly flexible tone control. You can lift the treble, the bass, or—and here is the unique feature—the middle frequencies to

Guidance Service, is available dally, including Saturdays, from 10 a.m. to 6 p.m., or will deal with enquiries by return of post. Our new illustrated Catalogue and Supplement will be a great boon to those desiring high quality equipment for modest expenditure. Send two 24d. stamps for your copy now. It may well save you pounds!

BSB MONARCH 3-SPEED MIXER CHANGER. New version with Acos Hi-g turnover pickup head with two separate sapphire styli £13/10/-. Portable cabinet 65/-, plus 5/- carr.

Leaflets on any of the above gramophone units free on

suit your own ear characteristics and the record or radio programme being heard. It is thus possible to arrange the frequency-response of the amplifier to a curve equal and opposite to the resultant curve of the other items in the chain so that what finally registers in the brain is as per original. This flexibility of control is even more importhe chain so that what finally registers in the brain is asper original. This flexibility of control is even more important than the nominal linear response of the amplifier, as the pickup, speaker, etc., are not linear. Independent Scratch-Cut is also fitted and special negative-feedback circuit employed. The Amplifier can accommodate a wide variety of records from old 78s to new L.P.s. Input is for all types of pickup of 0.1 v. output or more and there is full provision (and power) for Radio Tuner. It is available to match 2/3 or 15 ohms speakers. Price: 11 gns. (carriage 5/-). Fitted in portable Steel Cabinet, 2 gns. extra.

No. 2 "SYMPHONY" AMPLIFIER as No. I but with 10-watt Push-pull triode output and triodes through-. Woden mains and output transformers and choke, provision and power for Tuner. Output tapped 3, 7.5 and 15 ohms. Competes with the most expensive amplifiers on the market yet costs only 16 gns. (carriage 5/-). Fitted in portable Steel Cabinet 2 gns. extra.

"SYMPHONY" AMPLIFIERS with REMOTE CONTROL. Both the above model Amplifiers are available with all controls on a separate Control Panel with up to 4ft. flexible cable which simply plugs into the amplifier. Enables the Amplifier proper to be sat in the bottom of a cabinet whilst the controls are mounted

bottom of a cabinet whilst the controls are mounted conveniently higher up. Extra cost 2 gns.

STUDIO "SYMPHONY" AND DECCA AMPLIFIERS, Models I and 2. These amplifiers possess all the facilities of the above standard models together with valve amplification stage and precise tone correction circuits (separate for Std. and L.P.) to match the Studio Type "P" or Transcription and the Decca XMS Magnetic Heads respectively. Prices: No. 1, 13 gns. No. 2, 18 gns. Carr, 5/-.

CURRENT GARRARD, COLLARO & B.S.R. PRODUCTS AVAILABLE FOR IMMEDIATE DELIVERY FROM STOCK AT PRESENT.

MODEL TA 3-speed unit, with plug-in turnover head Type G.C.2, £10/16/-, or with Acos HGP 33 or 37 heads, £10/14/-, or with two separate high fidelity Acos HGP35 heads, £12/17/-. Unit less heads, £8/11/-, post 2/6. Heads, 42/3 each, post 1/-.
MODEL TB as above, but with long pickup arm. Less

MODEL TB as above, but with long pickup arm. Leas, 18/11/-, post 2/6.
Heads, 16/11/-, post 2/6.
Heads to fit this unit: Decca XMS, 54/6, Decca Crystal, 30/-, Acos HGP 55 42/3, Garrard Standard Magnetic, 28/-, miniature magnetic low impedance, 28/-, miniature magnetic high impedance, 38/-. Post on heads 1/-.

MODEL RC80M AUTOCHANGER. We recommend this as being the most mechanically perfect Autochanger on the market, and with absolute minimum motor noise—approaching Transcription quality. Price LESS HEADS £15/5/-. Price with short pickup arm and Garrard GC2 or Acos HGP37 turnover pickup Head £17/7/6 or with full-length Decca arm and complete with two Decca XMS Heads £20/15/- or with two Decca crystal Heads £18/10/- or two Acos HGP55 Heads £19/10/-. Stylus pressure accurately adjusted before despatch. despatch.

COLLARO PICKUPS AND HEADS. Studio Pickup Arm, 13/10. Studio Pickup head type "O" or "P," £3/0/9. Pickup complete £3/14/7. Studio Transcription Pickup Arm with Studio "P" head, £4/15/9. Ditto with Transcription head, £5/2/5.

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GARRARD Model 301, £25/3/6.

Cabinets available to house any of the above three motors together with pickup, price £3/7/6. Carriage 5/-.

"Symphony" No. 1 F.M. TUNER

We are pleased to announce that after extended research our new High Fidelity F.M. TUNER has been placed on the market and is available for immediate delivery from stock. It incorporates the latest type of permeability-tuned coil assembly of advanced design housed in die-cast protective anti-radiation shroud. The Tuner employs the most modern types of valves newly developed especially for F.M. circuits -ECC85, 2 x EF89, EB91.

The efficiency of the general circuit ensures extreme sensitivity and a very high music-noise ratio. The output impedance is ½ Megohm, rendering it suitable for feeding into any normal amplifier including those of the Highest Fidelity class.

A volume control is incorporated to adjust for variations in signal strength and amplifier input sensitivety.

radio/gram, selector switch and pickup socket are also incorporated, and the unit is readily linked to an A.M. Tuner without external changeover switch. The slow-motion tuning drive is especially smooth and free from backlash and the glass dial is illuminated. Overall size is: 9-in, wide x 6-in, deep x 6-in, high. Dial cut-out: 3-in, wide x 2-in, high.

The power requirements are: 6.3 v. at 2 maps, and 250 v. at 40 mA. Our model FMI Power Pack is ideal for providing this power and has capacity for the average A.M. Tuner as well.

The price of this high grade F.M. Tuner is only 14 gns. tax paid, and the Power Pack £3/7/6 extra if required.

SYMPHONY" BASS REFLEX CABINET KITS. 30in, high, consist of fully-cut \$\frac{1}{2}nt, chick, heavy, inert, non-resonant patent acoustic board, deflector plate, felt, all screws, etc., and full instructions, \$\frac{3}{2}in. speaker model, \$\frac{5}{2}in. speaker speaker. in our own laboratory and is your safeguard of optimum acoustic results. Carriage 7/6. Ready built, 10/6 extra.

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CONSOLE AMPLIFIER CABINETS. 33in. high, lift-up lid with piano hinge, take Tape Deck, Gram Unit or Auto-changer, Amplifier, Pre-Amplifier and Radio Feeder Unit finished medium walnut veneer. De Luxe version, price 11 gns. Oak or mahogany veneers 10/-extra. Special finishes to order. Carriage according to area, we will quote by return.

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No. 2/VS "SYMPHONY" SUPERHET TUNER. As No. 2 but incorporating on the wave-change switch an extra position for radio, giving T.R.F. bandwidth. Price 13 gns. Carr. and pkg. 5/-.

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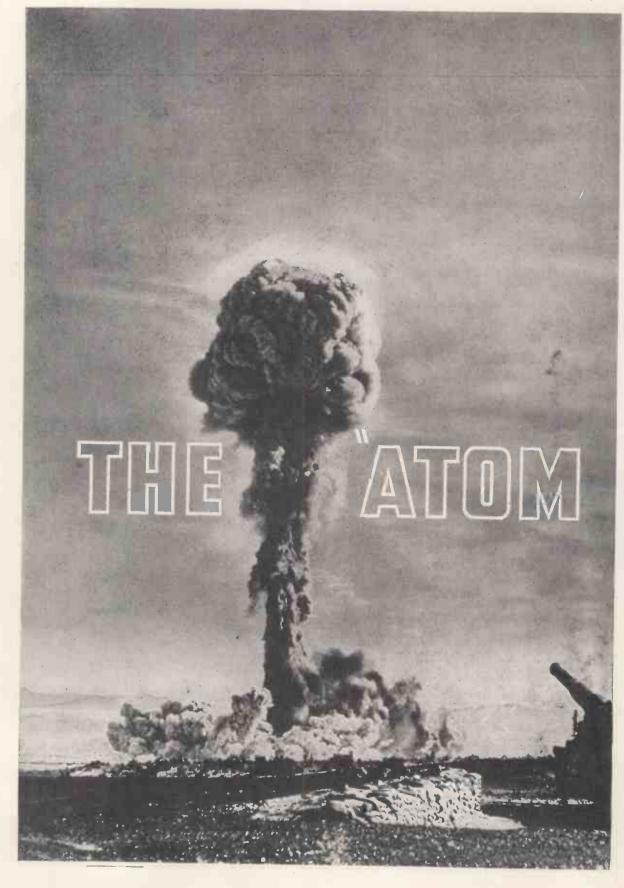
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TAPE AMPLIFIER TYPE C, expressly designed by Truvox to work perfectly with their Deck, 3 valves plus rectifier and Magic Eye level indicator. Price 16gns. Eye level indicator. Price 16gns. PORTABLE TAPE RECORDER CABINET to house Truvox Tape Deck and Amplifier together with speaker. Very strongly made and attractively finished in RexIne. Price £4/115/-, carriage paid. BRENELL TAPE DECK
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RECORD PLAYERS We are pleased to announce the entry on to the market of two "Symphony" Record Players designed to represent the greatest value in this line ever offered Model No. 1 contains the Collaro Model No. I contains the Collaro 3-speed single record playing unit AC3/554 and model No. 2 con-tains the Collaro Autochanger RC54. They are available with either Type "O" insert, "P" insert or transcription insert. insert or transcription insert. Prices (in attractive rexine case), No. 1 £10/19/6. No. 2 £14/19/6, Carr. 7/6. Transcription insert Carr. 7/6.



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Operating range, vehicle to vehicle, 30-50 miles in normal terrain.

Size and shape similar to the Pye W.S.19.

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Designed and manufactured by the designers of the W.S.19.

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10 kc/s to 300 mc/s

INPUT IMPEDANCE OF EACH PROBE:

Approx. 1 pf (input element of variable capacity divider)

MAXIMUM SENSITIVITY:

Full Scale Deflection for 1 Volt input

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Variable from .05 microsecs to 5 microsecs

RECURRENCE RATE OF MONITORED WAVEFORM: 100 c/s to 10 kc/s

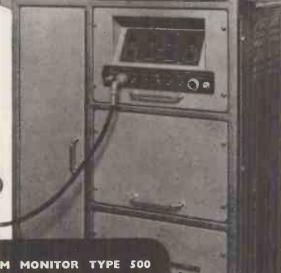
CALIBRATION :

Provision is made for accurate measurement of time and voltage scales of a waveform

PREVENTION OF JITTER:

A circuit is incorporated for providing a stable display when a monitored waveform is jittering with respect to its driving pulse.





HIGH SPEED RECURRENT WAVEFORM MONITOR TYPE 500

The wide bandwidth and high sensitivity of the instrument as well as the very high input impedance result from the use of a sampling technique. During each recurrence a measurement is made of the instantaneous amplitude of one point in the waveform. This measurement is amplified and applied to the cathode ray tube as one co-ordinate of a graph of the waveform. During subsequent recurrences, instantaneous measurements are made of different points, resulting, after about 100 recurrences, in a complete graph

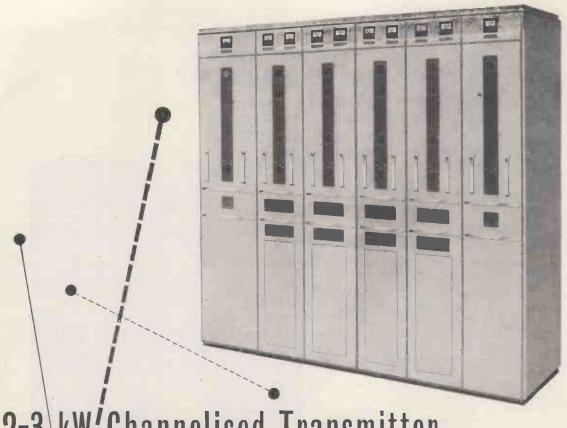
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2-3 kW/Channelised Transmitter

The GFT.560/2 is a 2-3 kW channelised transmitter with a frequency range of 1.5-30 Mc/s. It consists of three basic cabinets-r.f. unit, modulator unit, and power supply unit-combinations of which can be used to provide multi-frequency working as well as a number of different types of emission. The wave change facilities of the transmitter are both rapid and reliable—a valuable asset when the operating frequency is changed many times each day. The GFT.560/2 is fully tropicalised, and its unit construction facilitates future expansion of the initial installation, should the need arise.

For use in conjunction with the GFT.560/2 there are ancillary units that enable the transmitter to be remotely controlled over a two wire telephone circuit; operational adjustments are dialled to the transmitter.

The versatility and reliability of this new Mullard transmitter make it particularly suitable for h.f. en-route, ground-to-air services and point-to-point communication networks. A team of Mullard communication engineers is available to advise on the use of the GFT.560/2 in such applications.

> ABRIDGED DATA Frequency Range 1.5-30 Mc/s Frequency Stability To Atlantic City 1947 standards Power Output 3kW. c.w., 2kW m.c.w. or r/t Types of Emission c.w., m.c.w., telephony, frequency shift A,1, A2, A3, F1 Output impedance 600 ohms balanced twin feeder Power Supply 400V, 50-60 c/s, 3-phase.



CATHODEON

Quartz Crystals RELIABLE FREQUENCY

Prompt Delivery, all types, 2,000-20,000 kc/s

When ordering 10X replacements, why not use our hermetically sealed Type 2XL?



Garrard TRANSCRIPTION MOTOR

Features that the enthusiast will appreciate are the suppression of switch clicks, the extra heavy balanced turntable, and the very fine degrees of speed control available. Each of the nominal speeds, 78, 45 and 33½ r.p.m. can be adjusted by approximately 2½%. Wow and Flutter have been reduced to less than 0.2% and less than 0.05% respectively. The model is equipped for dual voltage ranges of 100 to 130 and 200 to 250 volts, 50 or 60 cycles according to the motor pulley fitted. The 301 is finished in quality grey tone enamel, is fully tropicalised and is supplied complete with plastic stroboscope, special grease, all fixing screws, washers, template and instruction manual.

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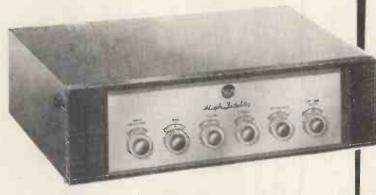
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A.78. 14 mv input for rated output. 500 c/s Turnover. 16 DB Roll-off at 10 K c/s.

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R.I.A.A. 11.5 mv input for rated output. 500 c/s Turnover. 14 DB Roll-off at 10 K c/s. 3 DB Flattening at 50 c/s to + 16 DB. at 30 c/s + 18.5 DB.

Crystal Pickup .35 volt with inbuilt equalisation from constant amplitude output to constant velocity output enabling switched replaying characteristics to be accurately employed.

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Microphone 6.5 mv for rated output. Flat characteristic.

Mixer Facilities for microphone input with radio/tape/gramo inputs.

Output. 1.2 volts from cathode follower stage.

Tape Recording Output. 1.2 volts cathode follower independent of monitoring.

Bass & Treble. Plus and minus 14 DB at 50 c/s and 10,000 c/s.

Volume. Twin ganged control giving correct gradation.

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Filter Slope. Variable to 35 DB per octave. Power Requirements. 375 v/7 ma. 6.3 v/l amp.

Price £48 0 0 complete



- * 12 watts undistorted output
- * Infinite damping factor



- ★ Frequency response substantially flat from 2 to 160,000 cycles
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- ★ Handles records, tape, radio and microphone

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PF 9113½" × 10" × 7". 34·5 × 25·4 × 17·7 cms.

PF 91 A 10\(^3\)'' \times 3\(^3\)'' \times 4\''. 25.6 \times 9.2 \times 10.2 \text{ cms.}

Master controls give extreme flexibility

SELECTOR

Six positions—tape, radio, microphone and correction for all types of records.

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10 indicated stages (5 cut, 5 boost), over 27 dbs. range. No distor-

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10 indicated stages (5 cut, 5 boost), over 27 dbs. range. No distor-

FILTER

Guts treble at 4,000, 7,000 and 12,000 cycles. This non-distorting device enables surface noise, high frequency interference, heterodyne whistle, etc., to be re-

VOLUME

Continuous volume control and on/off switch.



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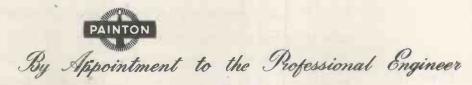
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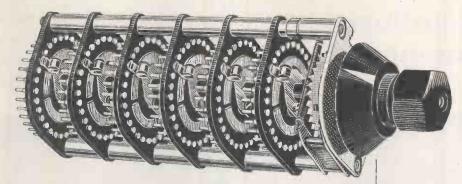
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PAINTON WINKLER SWITCH

VOLTAGE RATING: 250 volts A.C. / D.C. (maximum).

CURRENT RATING: 0.5 amp. (maximum).

Switching up to 29 positions (single-pole) per bank, or up to 30 positions per bank for 360° rotation.

SINGLE, DOUBLE, THREE-POLE or FOUR-POLE.

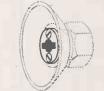
Painton Winkler Switches can be supplied for either 'Make-before-Break' or 'Break-before-Make' operation.

1-6 BANKS OPERATED FROM A COMMON SHAFT.

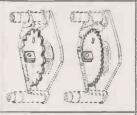
Each switch has an adjustable stop device, by which the switch can be set to the number of positions required.

AVERAGE CONTACT RESISTANCE : BETTER THAN 0,004 OHMS.

The distinctive Painton, knob type K21, with the 'adjustable skirt' feature has been specially designed to operate Painton Winkler Switches.



The white pointer can easily be lined up with dial markings. The friction-plate can be loosened by two screws, allowing the skirt of the knob to rotate.



Break-before-Make or Make-before-Break operation.

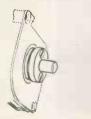


The 'direct-link' wiper provides a low capacity and inductance connection between the individual contact studs and the collector ring, and because the wiper is freely pivoted a constant and even contact pressure is obtained.



The contact studs are moulded into the nylon-filled phenolic resin panel, and though normally Silver-plated, can be specially Rhodium-plated if, required. The rigid stems of the contact studs are tinned to facilitate soldering connections.





The number of operating positions can be altered. Two stop plates can be adjusted by loosening a friction-plate clamped by two screws,

for the most <u>uniform</u> response

Of all the different bases that are used for magnetic recording tapes, none can match the precise uniformity of cast cellulose acetate. 'Scotch Boy III', with its cellulose acetate base, offers recordists the most exact uniformity of response that any tape can provide. 'Scotch Boy III' is the best of all tapes for high-precision recording, whether of voice, instrument, or mechanical sound.

For laboratory experiments that require the utmost uniformity of response 'Scotch Boy 111' is the natural choice: at 1000 c/s its

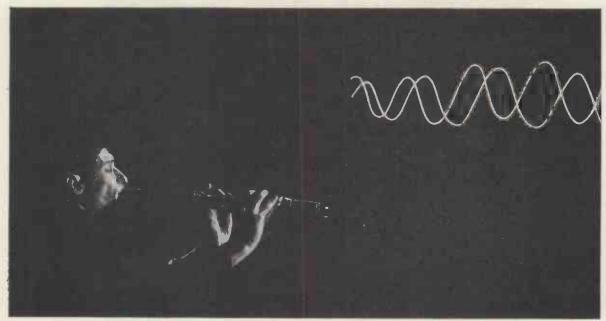


Photo and oscillograph of Cy Laurie playing a characteristically agile embroidery of a phrase from "King of the Zulus".

output variation within each reel is less than $\pm \frac{1}{4}$ db., and the variation from reel to reel is less than $\frac{1}{2}$ db. 'Scotch Boy III' is used by the services for experiments that involve the precise measurement of mechanical and other sounds, and by sound technicians and expert recordists all over the world.

'Scotch Boy 111' is supplied in 1200-ft. lengths on easily-threaded 7' plastic spools, and also in 600-ft. and 2400-ft. lengths. All these lengths are free from splices.



Record on 'Scotch Boy III'

-the tape with the cellulose acetate base

SCOTCH BOY'

MAGNETIC RECORDING TAPE

ANOTHER



PRODUCT

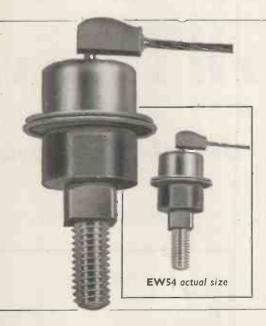
9.6.C.

EW54 Germanium Junction Power Diode

NOW AVAILABLE TO ELECTRONIC EQUIPMENT MANUFACTURERS

The EW54 p-n junction germanium diode is intended for use in rectifier circuits at medium voltage and current. The diode is hermetically sealed in a copper container. This is particularly important because of the deleterious effects of moisture on germanium devices.

The main features of this type of diode are high rectification efficiency and small size; the former results primarily from the very low forward resistance of the diode. The diode is of value in applications requiring outputs up to the order of 20A at voltages up to 50 (at 20°C), using a full-wave bridge arrangement.



For a typical diode, at an ambient temperature of 20°C:

Current at +0.5V=8Aat -100V=6mA

Other recent G.E.C. Semi-Conductors are the

EW53 EW58 EW59

Details of all the above devices may be obtained from

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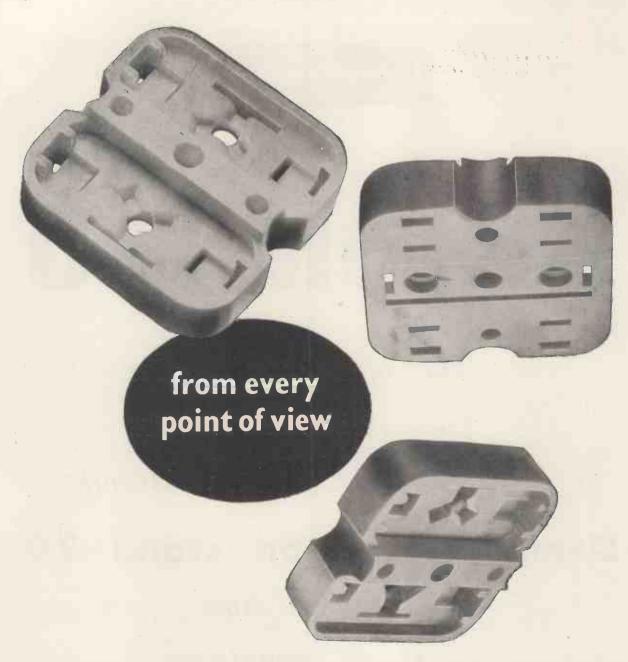
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AUGUST 1955

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Dear Sirs,
Please explain the best way to connect an external aerial to my portable s'het.
The outside aerial may be connected at either end of the F.A. or inductively coupled by several turns of the lead-in (not actually connected).



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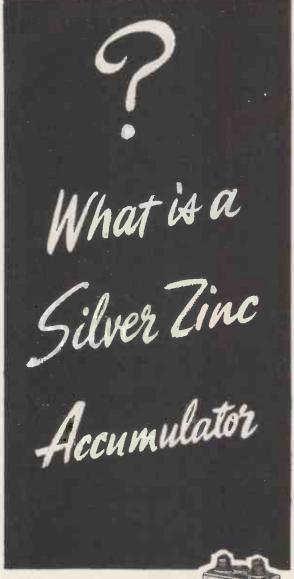
The illustration above shows the operator 'balancing the coil' of a "SCALAMP" GALVANOMETER — one of the final operations in its construction. This instrument, available in a range of sensitivities and incorporating a built-in four-position shunt, is one of the most versatile instruments of its kind. Housed in a robust dust-proof plastic case, it can be operated from mains or battery supply. Two new features, namely a slow-motion drive to the zero control, and an automatic self-shorting foot have recently been incorporated. The truly functional design of the case with built-in lamphouse, transformer and scale renders it adaptable to other instrument types. A FLUXMETER of exceptional performance and a series of ELECTROSTATIC VOLTMETERS (ranges 1-5, 3-10 and 5-18 kV) have already been introduced and a MEGOHM-PER-VOLT VOLTMETER is under development.

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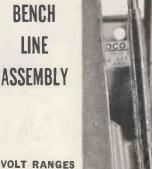
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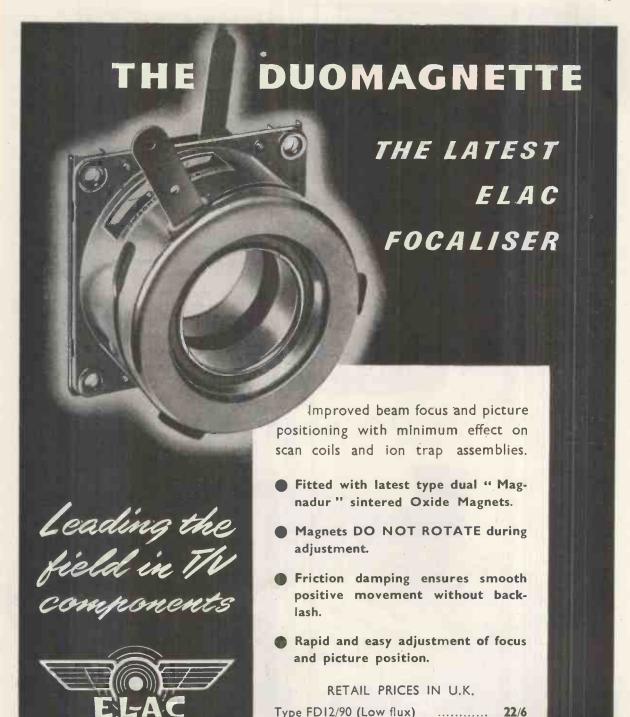
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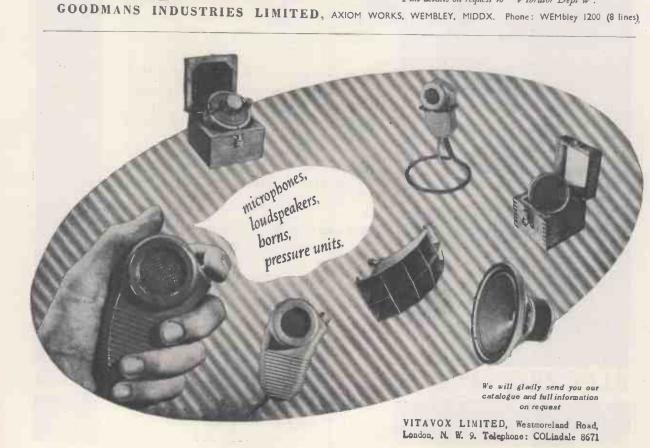
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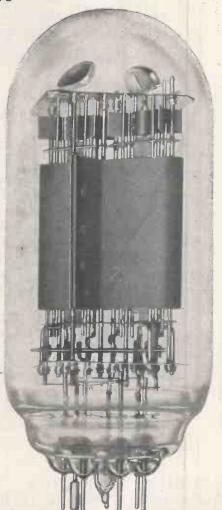
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Vg2 max	300 V		Pin 3	gl
VgI max	-100 V		Pin 4	gl k
Wa max	90 W		Pin 5	g2 a
Wg2 max	10 W		Pin 6	a
lk Max	800 mA		Pin 7	h



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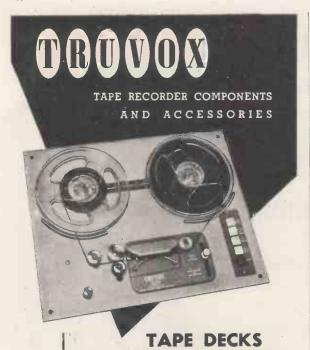
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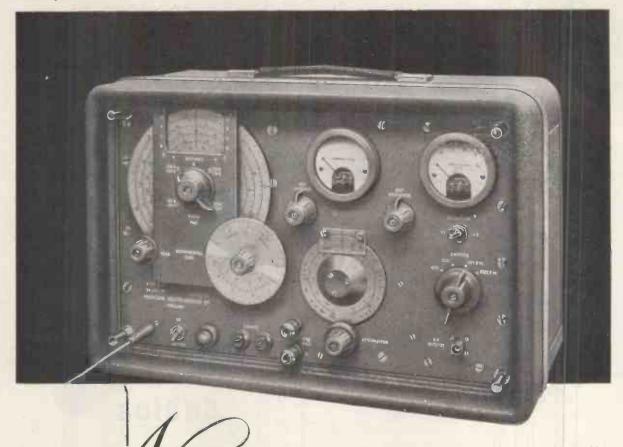
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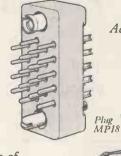
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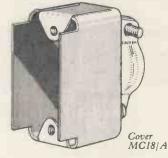
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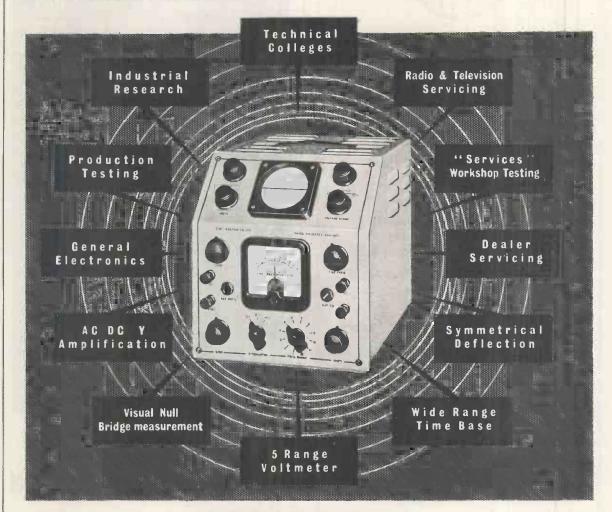
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THE McMURDO INSTRUMENT COMPANY LTD., Victoria Works, Ashtead, Surrey. Tel.: Ashtead 3401.

F. M.

NEW MINIATURE MEASURING OSCILLOSCOPE TYPE W.M.I.



SPECIFICATION

ELECTRICAL

Supplies: 110, 120 and 200 to 250 volts A.C. 50/60 c/s. (200-250 volts in 10V steps) 90 watts. Cathode Ray Tube: 21" (6.35 cms.) diameter, green phosphor, medium persistence. Voltmeter: Centre zero reading with 4" scale. Voltage Measurement: From 0.2 volt to 500 volts A.C./D.C., positive or negative, in 5 ranges. Y Amplifier: Single stage push-pull. D.C. coupled. Deflection Sensitivity: 1cm/volt. Bandwidth: 1-3 Mc/s. Input Circuit: D.C. or A.C. coupled. Maximum input: 500 volts A.C./D.C., positive or negative. Attenuators; 2:1, 10:1, frequency compensated Imput Shunt Resistance: 1 megohm. External reference

potential: Up to 500 volts, positive or negative. Time Base: Internal or External. Repetition frequency: 3 c/s to 120 kc/s in 12 ranges. Horizontal sweep speed: 0.025 cm/milli-second to 0.5 cm/micro-second. Synchronisation: External or Internal. Saw toeth waveform available for external use, amplitude 160 to 200 volts.

PHYSICAL.

Dimensions: Height 9½" (23 cms.); Width 8½" (20.5 cms.); Depth 11½" (29 cms.). Weight: 22½ lbs. (9.80 kgs.). Finish: Silver Grey Hammer. PRICE £49

For further details apply to:

E·M·I ELECTRONICS LTD

HAYES · MIDDLESEX · ENGLAND · TELEPHONE SOUTHALL 2468 (Ext. 655, 857 & 858)



EE 3



HARTLEY-TURNER SOUND EQUIPMENT

Introducing

THE HARTLEY-TURNER "315" LOUDSPEAKER

Here is the latest product from H. A. HARTLEY COMPANY LIMITED. After many months of tests we are now proud to announce our new "315" Loudspeaker.

This is a 12in. diameter unit with a frequency range from 20 c/s to 15 Kc/s obtained from a special coil and cone assembly. The voice coil impedance is 4 ohms and the power handling capacity is 15 watts. The frequency response is substantially flat over the range 30 c/s—18 Kc/s.

The speaker employs a die cast frame and is manufactured to the same exacting specifications that have made Hartley-Turner products renowned throughout the world.

Modern methods of production enable us to offer this new speaker

10 GUINEAS

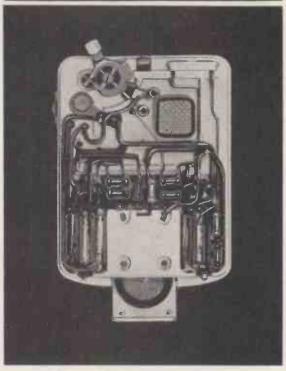
Full details sent free on request to:

H. A. HARTLEY CO. LTD. 152, HAMMERSMITH ROAD, HAMMERSMITH, LONDON, W.6.

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in equipment for the DEAF
and for PHYSIOTHERAPY



Actual Size of the A.V.C. Instrument (Covers removed).

AUTOMATIC VOLUME COMPRESSION

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Transistor "MINUET" Hearing Aid

Multitone were the first to introduce A.V.C. in Hearing Aids in 1936. It has proved essential for:—

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With the Minuet, two degrees of compression can be selected by the user. In addition the instrument is adjusted during fitting to take variations in individual optimum volume requirements.

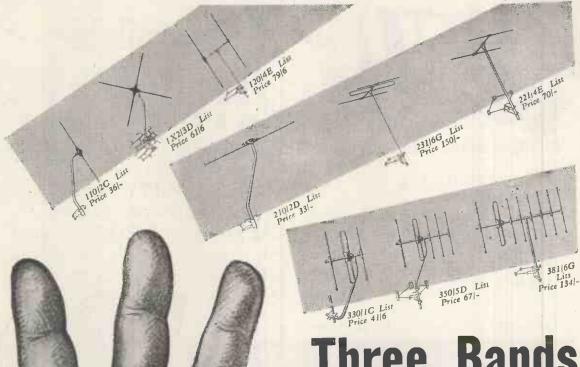
Weighs only 12 ounces complete with battery

Inquiries should be addressed to

MULTITONE ELECTRIC CO. LTD.

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hree Bands in hand

- Lower in price
 - Rods and insulators completely preassembled and aligned for peak performance.
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The QUALITY tape at a competitive price... 32/6 per 1200 ft reel

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WORKS & LABORATORY: 25 DASHWOOD TRADING ESTATE

LARCH ROAD · LONDON · SW 12

NEW SUPER LIGHTWEIGHT PICKUP MARK II giving an extended frequency range on L.P. disc.

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A. R. SUGDEN & CO. (Engineers) LTD. WELL GREEN LANE: BRIGHOUSE: YORKSHIRE Phone: Halifax 69169. Grams: Connoiseur, Brighouse

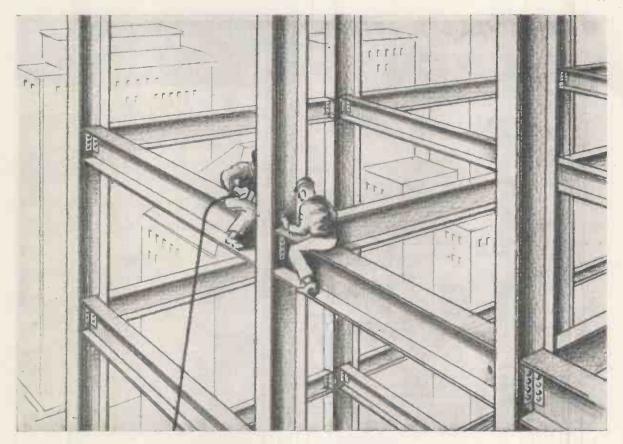
We present an entirely new three-speed unit operating at 334, 45 and 78 r.p.m. The full 12in. turntable is latheturned and manufactured of non-ferrous material. The main spindle is precision ground and lapped to mirror finish and runs in phosphor bronze bearings. The synchronous motor is dynamically balanced and resiliently mounted, making it virtually vibrationless, with low noise level and low hum induction.

The speed change is arranged mechanically and gives a 2% variation on all speeds, the synchronous motor running at constant speed at all settings. No braking action is employed to obtain speed change.

It is suitable for playing standard transcription and microgroove recordings. Input voltages 200/250 v. A.C. 50 cycles or, as specified to order for 200/250 v. A.C. 60 cycles, or 110 v. A.C. 50 or 60 cycles, Mounted on \$\frac{1}{2}\$ in. x 13\$\frac{1}{2}\$ in. with 3\$\frac{2}{2}\$ in. clearance distance below motorboard. Speed selector turret is fitted at left rear of motorboard. On-off switch at left front also releases pressure on the rubber drive assembly. All motorboards are drilled to take Connoisseur Standard and Super Lightweight Pickups unless otherwise ordered. When used with these pickups mounted in position, 3\$\frac{1}{2}\$ in. clearance above motorboard is recommended.

Price: £20 + P.T. £7 2s. 6d. Total £27 2s. 6d.

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Realism demands, in particular, the faithful reproduction of transients—one of the many merits of R. & A. reproducers.

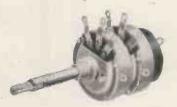


WHY ENGINEERS SPECIFY

EGEN

potentiometers-

Egen Potentiometers are based on long experience of requirements of television and electronic equipment manufacturers. In design, dependability, accuracy and freedom from wear they are *outstanding*, but, above all, they are completely NOISELESS.



DUAL POTENTIOMETERS
with concentrio operating
spindles. The new Egen
Dual Potentiometers incorporate all these outstanding design features
— multiple contact
rotors, smooth easy
movement, thorough
screening between sections, plus a convenient
soldering tag for earthing screened connec-

tions on each metal case. Switch and Potentiometer soldering tags are of high-grade brass heavily silver plated for easy soldering; they are positively located and withstand soldering heat and bending without loss of rigidity. Control spindles can be supplied to suit customers' requirements.

PRE-SET POTENTIOMETERS. Completely enclosed in high-grade phenolic mouldings. Solder tags heavily silver plated for quick soldering. Fully insulated spindles with integral control knobs. Tapped for 2-hole 6 B.A. fixing on \$\frac{4}{5}\text{centres.}\$ Type 126, wire-wound. Type 127, carbon.





STANDARD CARBON POTENTIO-METERS. Made by an entirely new method ensuring a highly stable resistance element, which is also very durable. Silent and smooth in operation, these controls offer both mechanical and electrical reliability. Soldering tags are heavily silver plated to resist oxidisation, and the mains switch has an efficient quick make-and-break action.

PRE-SET RESISTOR. This has a wire-wound resistance element, traversed by a nickel-silver slider. Adjustment is effected by a worm drive spindle fitted with a knurled and slotted knob. This

component is smooth and noiseless in action and is designed to meet the many and varied requirements of the Electronic Industry. Egen pre-set resistors can be supplied in multi-bank assemblies to suit individual requirements. There are also twin-track models, and types with an electrically divided slider, giving adjustment on two resistors with one operation.



EGEN ELECTRIC LTD. Charfleet Industrial Estate, Canvey Island, Essex • Phone: Canvey Island 691/2

MAINS TRANSFORMERS

FULLY	INTER	LEAVED
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SCREENED AND IMPREGNATED. ALL GUARANTEED

ALL PRIMARIES ARE 200/250 v. Half Shrouded.

	5 v. at 2 amps	16/3
ı	HS63. Output 250-0-250 v. 60 m/a., 6.3 v. at 3 amps., 5 v. at	
ĺ	2 amps	16/
IJ	HS40. Windings as above. 4 v. at 4 amps., 4 v. at 2 amps	16/
ì	Output,	
1	HS2. 250-0-250 v. 80 m/a	19/
ı	HS3. 350-0-350 v. 80 m/a. 19/ HS30. 300-0-300 v. 80 m/a	19/
i	HS2X. 250-0-250 v. 100 m/a., 21/-, HS75, 275-0-275 v.	
	100 m/a	21/
ı	HS30X. 300-0-300 v. 100 m/a., 21/ HS3X. 350-0-350 v.	
ı	100 m/a	21/
	,	

FSM63 (Midget). Output 250-0-250 v. 60 m/a., 6.3 v. at 3 amps.,

Fully Shrouded

F5 M63 (Midget). Output 250-0-250 V. 60 m/a., 6.3 V. at 3 amps.,	
5 v. 2 amps	16/9
Output	
FS2. 250-0-250 v. 80 m/a.	21/-
FS30. 300-0-300 v. 80 m/a, 21/-, FS3. 350-0-350 v. 80 m/a	21/-
F53V 3T0 03T0 100 1 20 F53, 73, 73, 75, 77, 77, 77, 100 11/2	23/-
FS2X. 250-0-250 v. 100 m/a. 23/ FS75, 275-0-275 v. 100 m/a.	23/-
FS30X. 300-0-300 v. 100 m/a. 23/ FS3X. 350-0-350 v.	
100 m/a.	23/-
All the above have 6.3 4-0 v. at 4 amps., 5-4-0 at 2 amps.	
FS43. Output 425-0-425 v. 200 m/a., 6.3 v. 4 amps., C.T. 6.3 v.	
4 amps., C.T. 5 v. 3 amps. Fully shrouded	47/6
FS50. Output 450-0-450 v. 250 m/a., 6.3 v. 2 amps., C.T. 6.3 v.	,,,
	4716
4 amps., C.T. 5 v. 3 amps. Fully shrouded	67/6
F35 X. Output 350-0-350 v. 250 m/a., 6.3 v. 6 amps., 4 v. 8 amps.,	
4 v. 3 amps., 0-2-6.3 v. 2 amps. Fully shrouded	65/-
FS160 X. Output 350-0-350 v. 160 m/a., 6.3 v. 6 amps., 6.3 v.	
3 amps., 5 v. 3 amps. Fully shrouded	44/-
FS43 X. Output 425-0-425 v. 250 m/a., 6.3 v. 6 amps., 6.3 v.	
6 amps., 5 v. 3 amps. Fully shrouded	63/6
HS6. Output 250-0-250 v. 100 m/a., 6.3 v. 6 amps., C.T. 5 v.	
Pise. Offput 250-0-250 V. Too mina., c.1. 5 V.	26/6
3 amps. For receiver R1355. Halfshrouded	20/0
HS150. Output 350-0-350 v. 150 m/a., 6.3 v. 3 amps., C.T. 5 v.	0-10
3 amps. Half shrouded	27/9
F36. Output 250-0-250 v. 100 m/a., 6.3 v. 6 amps., C.T. 5 v.	
3 amps. Fully shrouded	29/6
FS120. Output 350-0-350 v. 120 m/a., 6.3 v. 2 amps., C.T. 6.3 v.	
2 amps., C.T. 5 v. 3 amps. Fully shrouded	29/9
PRI/I. Output 230 v. at 30 m/a., 6.3 v. at 1.5/2 amps	21/-
FS150X. Output 350-0-350 v. at 150 m/a., 6.3 v. at 2 amps.,	
CT (2) and CT For and Company College Provided	31/6
C.T. 6.3 v. at 2 amps., C.T. 5 v. at 3 amps. Fully shrouded	21/0
The above have inputs of 200/250 v.	

OUTPUT TRANSFORMERS

MIDGET OP. 5,000Ω to 3Ω	3/9
OP10, 10/15 watts output. 20 ratios on Full and HalfPrimary	17/9 25/9
Williamson's O.P. Transformer to Author's specification £4/	
	32/-

FILAMENT TRANSFORMERS

All 200/250 v. Input.	
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@ 3 amps. F24. 24 v. tapped I2 v. @ 3 amps. F29. 0-2-4-5.6.3 v. @ 4 amps., I8/9. FUI2. 0-4-6.3 v. @ 3 amps. FU24. 0-12-24 v. @ 1 amp.	16/6 23/6 17/6 17/6
F5. 6.3 v. @ 10 amps. or 5 v. @ 10 amps., or 12.6 v. @ 5 amps., or 10 v. @ 5 amps. F6/4. Four windings at 6.3 v. tapped 5 v. @ 5 amps, each, giving	34/-
by suitable series and parallel connections up to 6.3 v. @ 20 amps.	51/6

Quotations, etc., stamped addressed envelope please.

C.W.O. (add 1/6 in £ for carriage).

Export enquiries invited.

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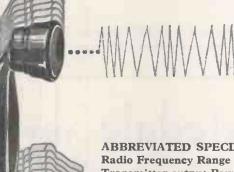
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RADIO TELEPHONE SYSTEM



This 7-channel Radio Link System has been designed for economy both in initial cost and maintenance demands.

This has been achieved without sacrifice of essential facilities or relaxation of performance standards. Both Radio and Carrier equipment for the 7-channel terminal is housed in a single 6-foot cabinet as illustrated. The equipment is fully tropicalized and suitable for continuous unattended operation in all parts of the world.



ABBREVIATED SPECIFICATION

Transmitter output Power

Baseband 7 Channels Maximum Deviation Receiver Bandwidth

60-216 mc/s 10 watts, or with Am-

plifier unit-50 watts

0.3-23.4 kc/s 50 kc/s

6 db down at ± 120 kc/s



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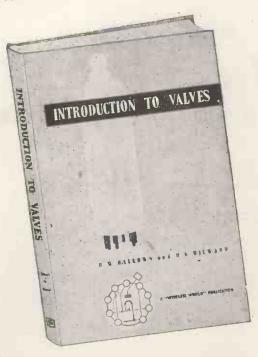
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ENGLAND



 $8\frac{3}{4}$ " x $5\frac{1}{2}$ " 152 pp. 107 illustrations 8s. 6d. net. By post 8s. 10d.

Describes the principles, operation and uses of radio valves

This book, by R. W. Hallows, M.A. (Cantab.), M.I.E.E., and H. K. Milward, B.Sc. (Lond.), A.M.I.E.E., describes the principles, construction, characteristics and uses of most types of radio valves. The approach is simple and as far as possible non-mathematical, but the book provides a thorough understanding of valves and how they work.

Published for "Wireless World"

Obtainable from all booksellers or from Iliffe & Sons Ltd., Dorset House, Stamford Street, London, S.E.1

Whartedale IMPROVED H.F. UNIT

SPECIFICATION

 Weight
 ...
 2 lbs. 13 ozs.

 Diameter
 ...
 $3\frac{5}{8}$ in.

 Cone Diameter
 ...
 $2\frac{3}{8}$ in.

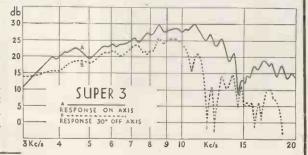
 Baffle Opening
 ...
 3in. diameter

Flux Density ... over 13,000 gauss
Total Flux 54,000

Power Handling Capacity 5/6 watts above 1000 c/s Weight of cone and coil assembly 1½ grammes Price £5 plus £1/13/3 Purchase Tax.

SPECIAL OFFER— During the months of September and October 1955 we are prepared to convert Super 5 units to the new Super 3 model at a nominal price of 15s. 0d. After this period the cost of conversion will be 30s. 0d. (Packing and carriage included.)

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3,000 C'S ON F.M. TRANSMISSIONS &
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£7.10.0 No. 150 (.150 in. dia.) for remote controls up to 6 in. in length

(For use without flexible casing)

The S. S. White Company will be pleased to advise which Outfit is most suitable for specific applications.

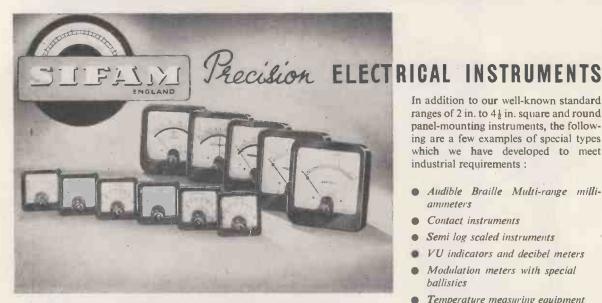
A detailed Parts List is available upon request.

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- Modulation meters with special hallistics
- Temperature measuring equipment
- Airborne instruments of special design
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57 GUINEAS

The heart of the matter . . .

The versatile F.C. 48 Chassis, hand-built by craftsnen, is fitted in both these fine Radlograms. The F.C. 48 has been specially designed to take the fullest advantage of modern High Fidelity recordings, and the unsurpassable quality of the new F.M. transmissions. This very latest ARMSTRONG 8-valve Chassis is the result of over 20 years of specialisation in high quality Radiogram Chassis. The Push-Puil Tetrode Output Stage will give more than 8 watts, with a Frequency Range of

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F.C. 48 CHASSIS £23. 18. 0

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These are a few of the towns where specially appointed dealers will be pleased to demonstrate this super instrument. Send a postcard for the name of your local dealer now.

All export inquiries to:

Messrs. Barnett Shipping & Export Co., Ltd., St. Magnus House, 25, Monument Street, London, E.C.3.

Manufactured entirely by:
Fisher Electronics Co., Ltd.,
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Protection against damage from IMPACT and VIBRATION







"BARRYMOUNT" cup-type isolators are designed primarily to absorb high-impact shocks with concurrent isolation of frequencies above 40 c.p.s. and general sound isolation. Utilisation of rubber in compression with substantially equal stiffness in all directions provides a smooth load-deflection curve.

Load ratings indicated for Mobile Applications (including shipboard installations) are such as to ensure a vertical natural frequency between 25 and 35 c.p.s. The design and assembly of the metal parts are such that they are self-captivating for maximum security.

Samples are available immediately ex stock

There are also air-damped types available for the protection of airborne equipment. "Barrymount" isolators are made in England under licence from Barry Controls Incorporated of U.S.A.



CEMENTATION (MUFFELITE) LTD., 39 VICTORIA STREET, LONDON, S.W.1



Cellular Polythene

INSULATED TELEVISION DOWNLEAD CABLES

To meet the exacting demands being made on the efficiency of aerial systems, the Glover range of Cellular Polythene insulated downleads have been designed to utilise the superior electrical properties of this new form of polythene.

Details of three designs are given as being most representative of modern practice.

The two Cables G.R.1., G.R.2. are intended for use in the service area and one G.R.3. for use in fringe areas and in situations where interference is high.

	CEL	LULAF	P	DLYTH	E N E.
Reference No.	G. R. 1. S	G. R. 1. F.	G. R. 2. S.	G. R. 2. F.	G. R. 3.
Characteristic Impedance ohms.	75	75	75	75	75
Service Area	LOCAL	LOCAL	LOCAL	LOCAL	FRINGE
Attenuation dB/100 fg. at 50 Mc/s.	3 · 0	3 · 4	2 · 3	3 · 6	1 . 5
200 Mc/s.				2 6	
Copper Conductor	1/-022	7/:0076"	1/:029"	7/:010"	1/:044"
Diam in inches:-				-,, 0.0	1) 044
Over Polythene.	0.093	0.093	0.128	0.128	0.200
" Wire Braid.	0.117	0:117	0.152	0.152	0.530
P.V.C. Sheath.	0.157	0.157	0.202	0.303	0.230



TRAFFORD PARK

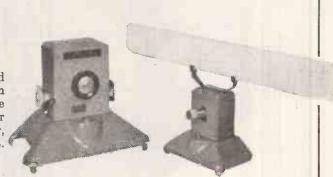
MANCHESTER 17 TRAFFORD PARK 2141



Laboratory Instruments

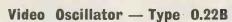
Torque Vane Wattmeter — Type U.181

A feed-through wattmeter for X Band waveguide No. 16. It will make an absolute measurement of power over the range 10 watts to 100 watts C.W., or the equivalent mean of pulsed power, with an accuracy of ±1%. The V.S.W.R. is better than 0.95.



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A wide range bridge for the measurement of resistance, capacitance and inductance between 15 Kc/s and 5 Mc/s. It will measure complex impedances balanced and unbalanced and between any pair of terminals in a three terminal network.



A portable instrument covering the range 10 Kc/s to 10 Mc/s with an output of + 10 dbs to — 50 dbs on 1 volt p. to p. amplitude stabilised to 0.5 dbs over its full frequency range. It includes a 50 cps. square wave output and facilities for direct reading of the modulus of the load impedance.



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Hi-Fi model with additional 10 in. speaker built in detachable lid

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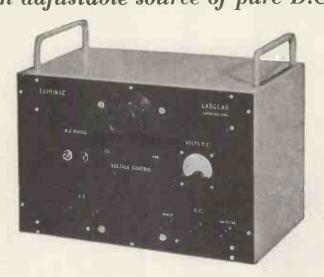
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S.L.86



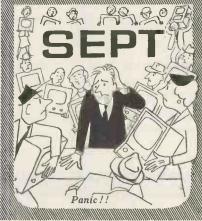


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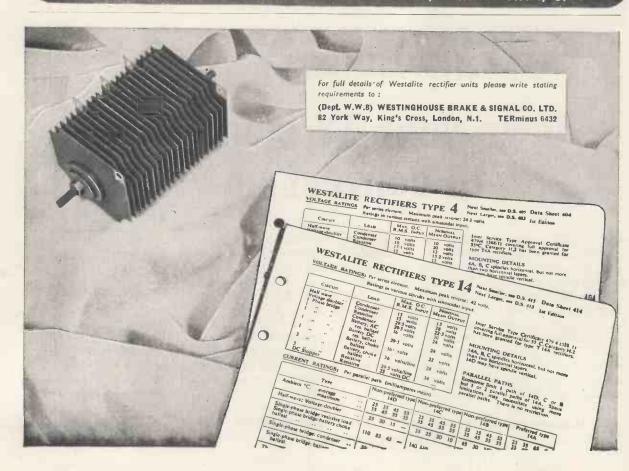


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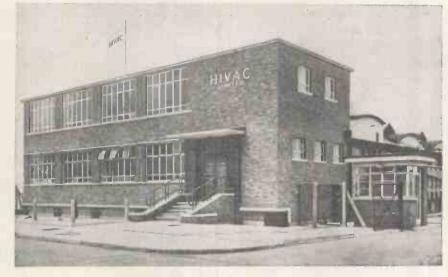
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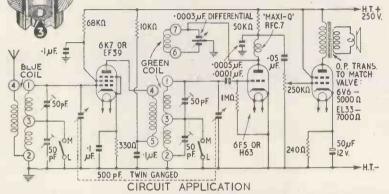
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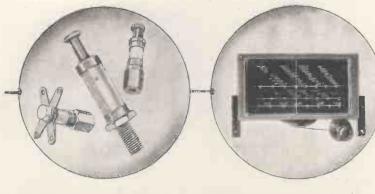
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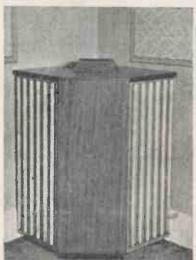


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Windings in parallel produce I volt 80 amps. A few at 28/6 (despatch 2/6).

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The finest bargain in first grade Time Switches.

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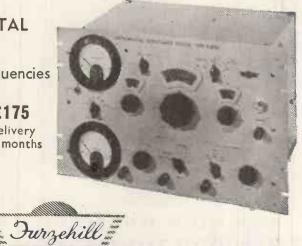
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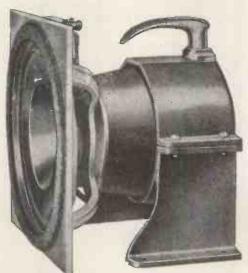
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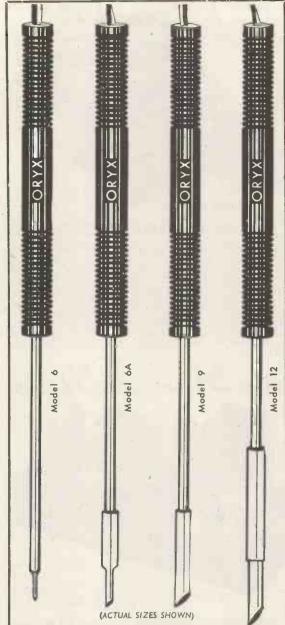
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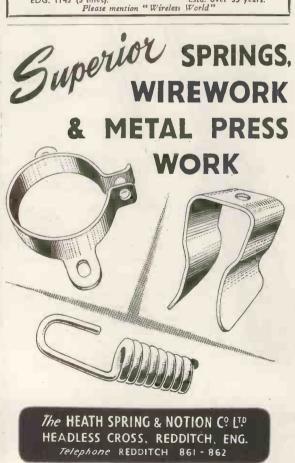
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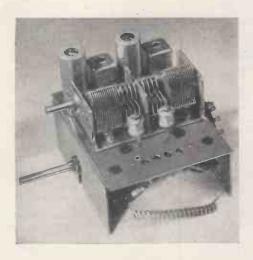
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6J5	5/-	50C5	10/-	ECL80	10/-	UY41	9/6
6J7	5/-	50L6	8/-	EF37A	10/6	V1507	9/-
6 K 7		72	4/6	EF39	6/6	VP4(7)	2/6
6K8		77	8/-	EF41	9/-		8/6
6L7		80	8/6	EF50(E)	5/-	VR105/30	9/-
6N7		83	8/-	EF54	5/-	VR150/30	8/6
6Q7		84/624	8/6	EF73	10/6	VT501	6/-
68 A		85A2	10/6	EF91	6/6	XH(I.5)	4/-
6SC		210LF	3/-	EL32	6/6	XSG1.5	6/6
6SG		2158G	4/-	EL41	12/6	X66	7/6
6SE	7 6/-	807	7/6	EL84	11/-	¥6 3	7/6
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All boxed and guaranteed. Post 6d. each. Immediate delivery. Full list S.A.E. Shop hours 8.30 to 5.30. Sats.
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for world wide reception

The Viscount Table Radiogram is a long range nine valve superhet with Garrard 3-speed Autochanger. Six electrically band-spread ranges in the 13, 16, 19-20, 25, 31 and 41 metre bands, M.S.W. and M.W. or M.W. and L.W., A.C. operation, push-pull output. Twin speakers. Large glass dial. 12 in. scale length each band. Separate bass and treble controls. Tuning indicator.

Reports reach the factory daily as to the excellence of its performance in all parts of the world.



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Full details on request.

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BRIGHOUSE

OSCILLOSCOPE MODEL 2300

This Model meets the need for a compact and robust instrument in which nothing has been sacrificed in order to achieve true portability. It has many of the facilities required in the laboratory as well as ruggedness demanded in the field. Note these features:

- Cathode Ray Tube diameter 23in. Square Wave Response adequate for Television synchronising wave-
- Direct-coupled X and Y Amplifiers.
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MAGNET WORKS

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B. H. MORRIS & CO. (RADIO) LTD. EST. 40 YRS

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THE NEW PREMIER TELEVISOR 13 CHANNEL DESIGN

SUITABLE FOR USE WITH ANY POPULAR WIDE ANGLE TUBE

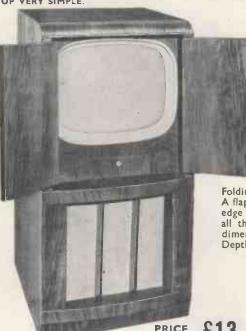
DESIGN I. Includes a Multi-Channel Tuner (Channels I-13) continuously variable 40 — 100 Mc/s and 170-225 Mc/s. The Tuner is supplied wired and tested and is complete with valves, all connecting leads and fixing

THIS DESIGN MAY BE BUILT FOR £34/9/7 (plus cost of

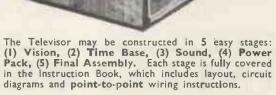
DESIGN 2. Channels I-5, tunable from 40-68 Mc/s. THIS DESIGN MAY BE BUILT FOR £30 (plus cost of C.R.T.).

- ★ Constructors who have built Design 2 (5 Channels) may convert their receivers to Design I for £6, this price includes Multi-Channel Tuner, New Vision Input Coil and full instructions.
- ★ All coils supplied for these two Superhet Receivers are PRE-TUNED ASSURING ACCURATE ALIGN-MENT and EXCELLENT BANDWIDTH.
- ★ Duomag permanent magnet focusing with simple picture centring adjustment.
- * Exceptionally good picture "hold" and interlace. Noise suppression on both Sound and Vision.

THE COMPLETE TELEVISOR IS SAFE TO HANDLE, BEING COMPLETELY ISOLATED FROM THE MAINS BY A DOUBLE WOUND MAINS TRANSFORMER. ALL PRESETICONTROLS CAN BE ADJUSTED FROM THE FRONT, MAKING SETTING UP VERY SIMPLE.



BAND 3 (COMMERCIAL)
For the Home Constructor



The Instruction Book also includes full details for converting existing Premier Magnetic Televisors for use with modern wide angle tubes. All components are individually priced.

Instruction book 3/6, Post Free. Includes details of both designs.

CONSOLE CABINETS

For 14", 16" and 17" Televisors

A handsome Walnut Cabinet that will be a fitting housing for a first-class Televisor.

Folding doors are fitted to cover the Cathode Ray Tube when not in use. A flap is provided which gives access to the preset controls on the front edge of the Chassis. A baffle board suitable for a 10in. Loudspeaker and all the necessary Tube and Chassis bearers are included. The overall dimensions of the Cabinets are the same: Height 38½in. Width 19in. Depth Top 19in. Depth Bottom 21in.

TUBE ESCUTCHEONS

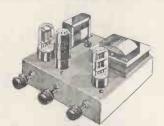
17in. White Moulded	21/-(pkg. & post 1/6)
17in. Bronze Moulded, complete with Protective Glass	48/- (pkg. & post 2/6)
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Rubber Ring (anti-Corona) for E.E.T.901	6/8
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PRICE £13-10-0 PLUS 21/- PKG. & CAR. 11 MONTHLY PAYMENTS OF \$1.0.0

TERMS OF BUSINESS: Cash with order or G.O.D. over £1. Please add 1/-for Post Orders under 10/-, 1/6 under 40/-, unless otherwise stated.

REMIER RADIO COMPANY

AMPLIFIER



MAY BE BUILT FOR

£4.10.0

Plus 2/6 Pkg. & Carr.

Vaive line-up 68L7, 6V6 and 6X5, FOR A.C. MAINS 200/250 VOLTS. The twin triode 68L7 is used for preamplification and also for a comprehensive tone control circuit, which includes two very wide range and continuously variable tone controls for bass and trebla. The output Vaive is of the beam type and feeds 4 watts into a specially designed output Transformer which is suitable for either 3 ohm or 15 ohm Speakers. Negative feed-back is applied from the secondary of the output Transformer over the whole Amplifier to the input stage giving an excellent frequency response. Due to the high gain and wide range tone controls any type of pick-up may be used. Overall size 9×7×5in. Price of Ampliher complete, tested and ready for use, £5/5/-, plus 3/6 pkg. and carr.

INSTRICTION BOOK. 1/- (Post Free) which includes

INSTRUCTION BOOK, 1/- (Post Free) which includes Assembly and wiring diagram, also a detailed Stock List of priced components.

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Electrostatic H.F. Speaker type LSH75 gain of 20 db over the range of 7-18 kcs., inherent capacity 800 pf., maximum polarising voltage 300 D.C., operative A.O. voltage 60, suitable for outputs of up to 5 watts, size 2\frac{1}{2}in.\times \frac{1}{2}in.\times \f

Type L8100, details as above, inherent capacity 1100 pf., suitable for outputs of up to 20 watts, size 5in. x 3in. x 3in. x price 21/-.
Will add sparkle to the top response of any equipment.

Type K3/25	E.H.T. Pencil Type S.T.C. 650 v. 1 mA	1/7
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K8/100		
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Type RM1		L/-
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	L.T. Type Full Wave	
12 v. 1 amp.		1/-
	1	//9
rs v. a amp.	18	0/6

A RANGE OF BAND 3 AND F.M. AERIALS IS NOW AVAILABLE

Teletron Ferrite Rod Aerials, Medium Wave 8/9. Medium/Long Wave 12/9.

Corner Reflex Cabinet for Goodmans Axiom 150 to specification £12/10/-, pkg. and carr. 15/-.

ALUMINIUM CHASSIS 18 s.w.g.

Substantially made from Bright Aluminium with four

7 x 5 1 x 2in		10 × 9 × 3in	7/-
7 x 3 l x 2 in	3/9	12 × 10 × 3in	7/9
94 × 44 × 2in	4/3	14×10×3in	7/11
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14 × 9 × 21 in. '	7/6		

ALUMINIUM PANELS IS S.W.O.

7 × 6in.	,	1/3	7 × 4in.	 1/-
94 × 6in.		1/8	9} × 4in.	
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12 x 9ln.		2/8	$12 \times 7 \text{in}$.	
14 × 9in.		3/2	14 × 7in.	
16 x 9in.		3/8	16 × 7in.	
20 x 9in.		4/8	20 x 7in.	 4/5
22 × 9in.		5/2	22 x 7in.	 4/11
20 x 9in. 22 x 9in.		5/2		

CABINETS-PORTABLE

Model PC/I Brown Regine co 15/11 Overall dimensions
15in, × 13 in, × 5in.
Clearance under
when closed 2 in.
Model PC/2 Grey Lizard Regine covered 45/-Overall dimensions 15in. × 13in. × 6in. Clearance under when closed 3in. lld Model PC/3 Rexine type covering in various cols., 69/6.

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16fin. × 14fin. × 10fin.
Clearance under lid when closed 6fin. All the above Cabinets are supplied with Panel, Carrying Handle and Clips.

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V.H.F. Tuning Unit type UT340 permeability tuned, coverage 86-103 mcs. stage gain Aerial to output of 1st I.F. (contained in Unit) approximately 350. Maximum frequency drift 0-70 degrees centigrade 30 kcs. Radiation less than 26 microvolts per metre, price 59/5 (including tax). Valve UCC85.

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Overall bandwidth of the above Units 200 kcs.

Complete Handbook containing full details of construc-tion and point-to-point wiring diagrams including also details of F.M. Aerials 2/6 post free.

*QUALITY CRYSTAL PICK-UP ROTHER-MEL TYPE U48 26/-. Plus 1/6 Pkg. and Carr.

The New

"PREMIER PORTABLE"

USING THE NEW LANE 2-SPEED TAPE UNIT MARK 6

COMPLETE

Packing & Carriage I gn.
(Including Reel of Scotch Boy Tape and Microphone)

H.P. Terms: Deposit £10/8/- and 12 monthly payments

of £2/16/-.
or Complete Kit Including All Parts, Valves, Speaker Cabinet, Tape Unit, Reel of Scotch Boy Tape, Rewind

Spool and Microphone at £37.4.0 plus pkg. & carr. 15/-. H.P. Terms: Deposit £9/7/9 and 12 monthly payments of £2/11/-.

SPECIFICATION

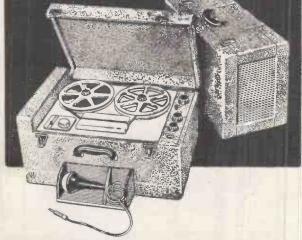
HIGH QUALITY

BASS CONTROLS.

ING MICROPHONE.

SPECIALLY DESIGNED MIC-ROPHONE BY A LEADING MANUFACTURER.

- ★ TWO SPEEDS 7½in. AND 3½in. ★ 7-VALVE E PER SECOND. ★ THREE SPECIALLY DESIGN- ★ INDEPENDENT TREBLE AND
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- 1,200FT. TAPE REELS PROVIDING PLAYING TIMES OF 1 HR. AND 2 HRS. * AMPLIFIER MAY BE USED FOR RECORD REPRODUCTION OF HIGH QUALITY.
- * DROP-IN TAPE LOADING.
- EASY FORWARD OR REWIND WITHOUT REMOVING TING MICROPHONE. TAPE.
- * ONE KNOB DECK OPERA-



SEPARATE UNITS CAN BE SUPPLIED AS LISTED BELOW: Amplifier (built, wired and tested with Speaker). £14/15/-, plus

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Amplifier Kit (including Speaker). £11/-/- plus packing and

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New Lane 2-speed Tape Unit Mark 6. £18/10/- plus packing and carriage 7/6.

Deposit £4/11/9 and 12 monthly payments

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Reel Scotch Boy Tape MC2-111 (1,200ft.), £1/15/-, plus packing and carriage 1/-.

Instruction Booklet. 2/6. Post free.

PREMIER RADIO COMPANY =

WILLIAMSON AMPLIFIER KIT 15 gns.

H.P. Terms: Deposit £3,18.9 & 12 monthly payments of £1,1.8 This Kit is absolutely complete and all components are guaranteed exactly to author's specification

WILLIAMSON OUTPUT TRANSFORMER Author's Specification 3.6 ohms secondaries

MAINS TRANSFORMER SP425A (Completely Shrouded)

This Transformer has an additional 6.3 v. 3 A. and is capable of supplying an extra 50 mA. for Pre-amp £2.12.6

WILLIAMSON CHOKES 30H 20 mA. Fully shrouded 11/9

Full Scale	External		
Deflection	Dimensions.	Movement	
	in.		
3.5 A	2½×2½	R.F. Thermo	7/6
20 A	2 round	M/C	8/6
40 A	21 round	M/C	8/6
500 mA.	24 round	M/C	10/6
30 A	21 × 21	M/C	8/6
50 mA.	2½×2½	M/C	7/6
20 V	2½×2½	M/C	6/6
1 mA	2} round	м/С	22/6

CRYSTAL MICROPHONE **INSERTS**

Ideal for tape recording and amplifier. No Matching transformer required, 8/6 post free



PREMIER MAINS TRANSFORMERS

All primaries are tapped for 200-230-250 v. mains 40-100 cycles. All primaries are screened. SP175B, 175-0-175, 50 mA., 4 v. @ 1 a., 4 v. @ 15/-2-3 a. SP350A, 350-0-350, 100 mA., 5 v. @ 2-3 a., 6.3 v. 21/-30/-30/-

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37/6 £3/7/6

Build these NEW PREMIER DESIGNS

3-BAND SUPERHET RECEIVER



MAY BE Plus 2/6 BUILT FOR £7.19.6 Pkg. & Carr.

BUILT FOR **21.13.0** Pkg. & Carr.

Latest type Superhet Circuit using 4 valves and metal rectifiers for operation on 200/250 volts A.C. mains. Waveband coverage — short 16-50 metres, medium 180-550 metres, and long 900-2,000 metres. Valve line-up-6K8 freq. changer, 6K7, IF, 6Q7, Detector AVC and first AF, 6V6 output. The attractive cabinet to house the Receiver size 12in. long, 6½in. high, 5½in. deep can be supplied in either WALNUT or IVORY BAKELITE or WOOD. Instruction Book. 1/- post free, which includes assembly and wiring diagrams, also a detailed stock list of priced components.

TRF RECEIVER



MAY BE BUILT FOR

£5.15.0

Plus 2/6 Pkg. & Carr.

The circuit is the latest type TRF using 3 valves and Metal Rectifiers for operation on 200/250 A.C. Metal Rectifiers for operation on 200250 A.C.
mains. Waveband coverage is 180-550 metres on
medium wave and 800-2,000 metres on long wave.
The dial is illuminated and the Valve line-up is
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The attractive Cabinets to house the Receiver size 12in. long, 64in. high, 54in. deep, can be supplied in eigher WALNUT or IVORY BAKELITE or WOOD.

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ALL-DRY PORTABLE RAD10 RECEIVER



£7.8.0 Plus 2/6 Pkg. MAY BE BUILT FOR

4 miniature Valves in a Superhet Circuit covering medium and long waves. Rexine covered Cabinets II in. x 10in. x Rexine covered Cabinets Il in. x 10in. x 5jin. in two contrasting colours. Wine with Grey Panel, or Blue with Grey Panel, please state choice when ordering. THE SET MAY BE USED EVERY-WHERE—home, office, car or holidays. INSTRUCTION BOOK. 1/6 (Post Free) which includes Assembly and wiring diagrams, also a detailed Stock List of priced components.

DECCA MODEL 33A RECORD PLAYER ADAPTABLE FOR STND. OR L.P.

ADAPIABLE FOR STND. OR L.P.

Includes crystal pick-tup with sapphire stylus and a light-weight plastic spring balanced arm. Heavy gauge pressed steel case with brown-enamel finish in good quality for operation on A.C. mains 200/250 v. 6.c.p.s. Supplied complete with single dead (either standard or long playing.) 24/19/6.

Extra Head can be supplied. Flus

Extra Head can be supplied. Plus Pkg. and Carr. 5/-.

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DECCA MODEL 37A. Appearance as above (Model 33A) lever speed change 33; r.p.m. and standard, crystal turnover Head. £6/19/6. Plus packing and carr. 5/-. LOUDSPEAKERS

ELAC ELIPTICAL 7" × 4" 21/10

ELAC-2jin. dia. Moving Coil, 15 ohm imp..... 15/-ELAC-8in, dia. Moving Coil 3 ohms imp.

PLESSEY—8in. dia., Mains Energised, 3 ohms imp. (600 ohms field) with Pentode Transformer 22/6

PLESSEY—10th. dta. storing Coil, 15 ohms.
GOODMANS—12th. dta., Moving Coil, 15 ohms.
£8/12/6

Plus 5/- packing ane carriage.
VITAVOX-K12/20 12ln. dia., Moving Coll £11/11/-

Plus 5/- packing and carriage.

85

PLESSEY 12"

3-SPEED AUTOMATIC RECORD CHANGER

Made by World-famous manufac-turer. The Unit designed to play 12lm., 10ln. and 7ln. Records intermixed in any order at 33; 45 or 78 r.p.m. Capa-city 10 records. New reversible dual stylus crystal Pick-up has extended frequency range. For use on frequency range. For use on 100/125-200/250 volts 50 cycles. A.C. mains. LIMITED QUANTITY ONLY.

Plus packing and carriage 5/-.
BRAND NEW, guaranteed and in manufacturers' original

£9.19.6 LIST PRICE

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DARK SCREEN FILTER IN TRIPLEX GLASS 16° x 18° SUITABLE FOR ALL TUBES UP TO 17° 10/-, PLUS PACKING & POSTAGE 1/6d. LATEST TYPE RUBBER ESCUTCHEON SUITABLE FOR 17 RECTANGULAR TUBES AT A SPECIAL PRICE OF 10/-, PLUS PACKING & POST, 1/6d.



ACCUMULATORS

volt 10 amp. (by famous maker)

MOVING COIL METER

A super quality Moving Coll Meter basic movement 2 m.A.
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2 in. Overall dimensions
2 in. Overall dimensions
2 in. Hencep. Bakelite Case projecting type, at
present scaled 1 amp. R.P. By removing thermocouple,
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ACOS. "MIC30" impedance Crystal Microphone £2/10/-.

(This Microphone can be used as either Hand or Deak

type.)

SPECIAL OFFER

Acos Microphone type 22-2 complete with Stand39/6 Plus Postage & Packing I/-

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Two specially designed chokes with three smoothing condensers with circuit diagrams. Cuts out all mains noise. Can be assembled inside existing receiver, 4/11, plus 6d. pkg. and carr.

Germanium Crystal Diodes. G.E.C. wire ended, 2/6. 24/- doz.

PREMIER RADIO COMPANY

1155 RECEIVER UNIT

NEW CONDI-



POWER SUPPLY UNIT WITH OUTPUT STAGE FOR ABOVE



Jones plugs for connecting the Power Pack to the Recciver are included. The 6V6 output stage complete with Output Transformer and 6in. speaker is built into the unit. Price £5/5/p plus 5/- packing and carriage. The two above units together on Hire Purchase Terms £4/4/6 deposit and 12 monthly payments of £1/3/10 plus 15/6 pkg. and carriage.

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SLIDER RESISTANCE. Geared adjustments, 7.5 ohms, 4 a., 12/6, postage and carriage 1/6.

HEAVY DUTY L.T. TRANSFORMER. Primary tapped 180-230 volts, 50 cycles. Secondaries 4.2 v. 10 a. 4.2 v. 10 a., 25/-, postage and carriage 2/6 ROTARY RESISTANCE. Wire-wound heavy duty 14 k. ohms, 7/6, postage and carriage 1/-.

MINIATURE TUNING CONDENSERS 6/9

2-gang .0005 mfd, with trimmers

PREMIER VARIABLE IMPEDANCE "MATCHMAKER" M.O.IS OUTPI TRANSFORMERS

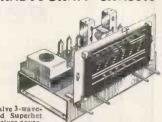
Designed to meet the demand for an efficient variable ratio Output Transformer 11 ratios from 13:1 to 80:1, all centre tapped and can be used to match any output valves either single or push-pull Class 'A', 'AB1', 'AB2' or 'B' to any low impedance speech coil or combination thereof. Primary Inductance 50 henries 15 watts audio 100 mA. Price 45f-.

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RADIOGRAM CHASSIS



5 Valve 3-wave-band Superhet
Receiver covering short, medium and long waves. Using the latest
miniature all glass valves, overall chassis size
13in. x 7in. high x 6in. deep, dial aperture 10in. x
4im. BRAND NEW, READY FOR USE AND
GUARANTEED.
Postage and packing 10j-.
Or on Hire Purchase terms, deposit £2/5j- and
9 monthly payments of £1.

CABINET available for above Chassis in figured walnut lined with white sycamore, size 3ft. wide, 2ft. 8in. high, ift. 5in. deep, £15/15/-. Or on Hire Purchase Terms, deposit £3/18/9 and 12 monthly payments of £1/1/8. Packing and Carriage extra

PORTABLE TAPE RECORDER CABINETS

All Rexine covered

Amplifier Туре Price Tape Deck Lane Mk. VI Truvox Mk. III Truvox Mk. III Truvox Mk. III Mk, V1 T.D.1 T.D.2 T.D.3 £4/19/6 £4/4/-£4/4/-£4/4/-Premier E.A.P. Premier Truvox C Plus Postage and Packing 5/-.

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Explains the characteristics of the cathode ray oscilloscope and its many uses in servicing radio receivers, amplifiers and television equipment. Special attention is given to the correct interpretation of oscillograms and the work is illustrated by practical examples of oscillograms and circuits.

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Wireless World

RADIO, ELECTRONICS, TELEVISION

Managing Editor:
HUGH S. POCOCK, M.I.E.E.
Editor:

H. F. SMITH

AUGUST 1955

In	This	Issue
	T 11663	133566

VOLUME 61 NO. 8 PRICE: TWO SHILLINGS

FORTY-FIFTH YEAR OF PUBLICATION

Editorial Comment	• •		353
Colour Television Standards			355
Short-wave Conditions			355
World of Wireless ,			356
Inexpensive Wave Analyser. By M. G. Scroggie			360
Letters to the Editor			367
Band II F.M. Tuner Unit. By L. Hampson			368
New Navigational Aids	- 1		3 75
The Automatic Factory	* *		378
Wide Range Electrostatic Loudspeakers—3. By P.	7. Wai	ker	381
High-Fidelity Home. By Richard Arbib			385
Transistor Equivalent Circuits—2. By W. T. Cocking			388
Transmitting Colour Information	••	• •	393
The Cascode. By "Cathode Ray"			397
Electronic Teleprinting			401
Random Radiations. By "Diallist"	• •		402
Unbiased. By "Free Grid"	*.*		404

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32. HIGH GAIN VIDEO AMPLIFIER

FIG. I

The gain of a video amplifier with a given bandwidth, using a high slope r.f. pentode, is limited by the output or load capacitance. A typical cathode compensated circuit using an EF80, with a bandwidth of 3Mc/s (6dB), a maximum video output voltage of 50V, and a load capacitance of 26pF, has a gain of 11. The gain can be increased and the bandwidth maintained if the effective load capacitance is reduced by means of a cathode follower. The practical circuit shown in Fig. 1 has a gain of 22. A Mullard PCF80 triode pentode combines the functions of video amplifier and cathode follower in one envelope.

0-8mA 3-7mA SI5kΩ IW VIDEO 3·5V

Video Amplifier

The cathode of the pentode (Fig. 1) is held at about + 3.5V by the pentode cathode current and the bleed current through the 27kΩ resistor. With the screen grid at 190V this gives an anode current of 3.5mA. The permissible anode current swing is limited by the maximum screen grid dissipation rating of 0.5W, therefore the output voltage is limited to about 80V. The dynamic characteristic of a typical PCF80 pentode shows that the corresponding maximum limit of the anode current is 8.8mA. The effective slope is 1.5mA/V, therefore the gain gmR1 = $1.5 \text{mA/V} \times 15 \text{k}\Omega = 22.5.$

Cathode Follower

Mullard

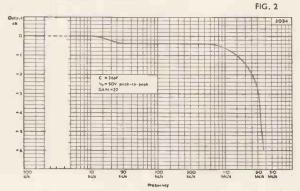
The total capacitive load of a video stage is made up, in a typical good design, approximately as follows: Video output 4pF, C.R.T. 8pF, Synchronising separator 6pF, Noise suppressor 2pF, Strays 6pF = 26pF.

If the total gain of the stage is to be greater than 20 and a cathode follower output (with a gain less than 1) is to be used, the amplifier must have a higher gain—say 25. To achieve this the output capacitance must be less than 11pF. With a cathode follower the capacitive load of the amplifier is reduced to: Video output 4pF, Cathode follower input 3pF, Strays 3pF = 10pF, which is within the 11pF limit. The overall response is limited by the cathode follower, which must be capable of developing an output voltage of at least 50V across a capacitive load of 26pF.

> Further notes on the circuit are available with reprints of this advertisement. Reprints of all advertisements in the Valves, Tubes and Circuits series may be obtained free from the address below.

Cathode Current Requirements

The cathode current has two components: a current which maintains the signal across the cathode resistor, and a transient current which charges the output capacitance. The valve must supply both currents. If it failed to do so the cathode voltage would not follow the grid voltage. If the cathode cannot rise at the same rate as the input the valve will run into grid current, and the positive going edge of the signal will be distorted. Alternatively, if the cathode cannot fall at the same rate as the input, the valve will be cut off and the negative going edge will be distorted. In general the second alternative is worse, as cut-off is an absolute limit to change in the cathode current. The onset of grid current does not preclude further change if the source impedance of the driver stage is reasonably low. As each input pulse has two transients associated with it the total current swing is the maintenance current plus twice the transient current. To prevent cut-off the standing cathode current must exceed the sum of the transient and maintenance currents.



Derivative Correction

It is possible to compensate for a poor fall time in a cathode follower by adding to the output a voltage proportional to the derivative of the video input signal. This voltage, which can be obtained by the inclusion of a resistor in the triode anode circuit, may be a.c. coupled to the c.r.t. grid when the cathode follower is driving the c.r.t. cathode. The derivative correcting voltage is thus added in the correct phase. It is found that this arrangement increases the contrast of the high definition bars of test card 'C' and improves the vertical edges; but it tends to produce ringing and 2.5Mc/s oscillation, therefore the use of derivative correction with this circuit is not recommended.

Performance

In a 21-inch tube receiver the recommended circuit (Fig. 1) gives a composite video output up to 80V before there is any visible sign of cathode follower overload. The frequency response, measured with a sine wave input and a 50V peak-to-peak output, is shown in Fig. 2. The stage gain is 22.

MULLARD LTD., Technical Service Dept., Century House, Shaftesbury Avenue, London, W.C.2



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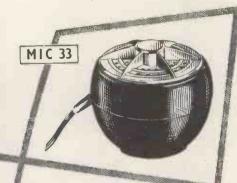
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A crystal hand or desk omnidirectional microphone for the high quality public address and tape recording field, incorporating a specially designed acoustic filter giving a response flat from 30 to 7,000 c/s.

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MIC 36

A handsome omni - directional instrument of high sensitivity and a substantially flat response from 30 to 7,000 c/s. Alternative models, with or without switch, are available with suitable adaptors for floor or table stands or for hand use.

RETAIL PRICE: £3 - 3 - od. without switch or £3 - 8 - od. with switch.

MIC 35

A general purpose hand microphone of robust construction with substantially flat response from 50 to 5,000 c/s. Suitable for recording apparatus. Public Address equipment etc. RETAIL PRICE fi - 5 - od.

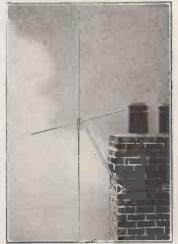
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"BELLING-LEE" NOTES

AERIALS FOR THE V.H.F. BROADCAST BAND II FREQUENCY MODULATED



The period of experimental B.B.C. transmissions from Wrotham has

passed.

From now onwards the Wrotham transmitter will send out three programmes, the Home, the Light and the Third, on Band II. These transmissions will be frequency modulated and will require horizontal aerials which are in the position of maximum response when broadside on to the transmitter. The B.B.C. V.H.F. programmes will eventually practically cover the country, and this will allow listeners to enjoy a quality of reception hitherto un-obtainable. The next transmitters to come into operation will be Pontop Pike (Newcastle) and Divis (Belfast) in that order. In many difficult locations listening will become a pleasure for the first time. New or modified receivers will be required and these will have built-in general purpose aerials. However, tentative field trials indicate that, in many cases, and certainly at distances in excess of 40 miles from the transmitter, particularly in hilly country, a better aerial will be required. In a number of cases a "Lofrod" loft aerial will be satisfactory but at these distances many listeners will require an outdoor aerial and in really difficult locations a 3-element or an horizontal "H" will be necessary. Above we illustrate V.H.F. stubs attached to a dipole. These may also be fixed to any of our aerials of the parasitic element type.

Advertisement of BELLING & LEE LTD. Great Cambridge Rd., Enfield, Middx. Written 20th June, 1985

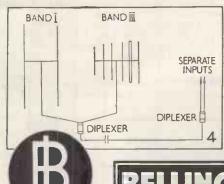
DIPLEXER TUNED FILTER

For use with band III aerials, adaptors, and combined band I/band III aerials



If a television receiver has a single input socket, and if separate band I and band III aerials are being used, viewers can either put up with the inconvenience of changing over the aerial every time they switch to the other programme or they can terminate both aerials into a diplexer and connect the socket output from this to the receiver input. Both signals are available at the receiver, but when the selector switch is tuned to one programme, the receiver does not have to contend with interference picked up on the temporarily unwanted aerial. If the feeders from the two aerials are just bunched and connected to the receiver input, there might be ghosting interference from both aerials instead of from only one, impaired signal strength due to mis-match, troubles due to standing waves on the feeders, or other interaction.

- A diplexer is necessary where separate band I and band III aerials are fed to a receiver having a single input for both frequencies.
- 2 A diplexer is not needed where separate band I and band III aerials are fed to a receiver having separate input sockets.
- 3 A diplexer is only required with combined band I/band III aerials, or aerials fitted with adaptor units, where the receiver has separate inputs.
- 4 On long cable runs, it is often more economical to fit a diplexer at each end, utilising a single feeder low-loss cable.



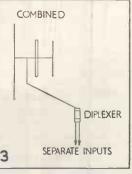
BANDI BANDII

DIPLEXER

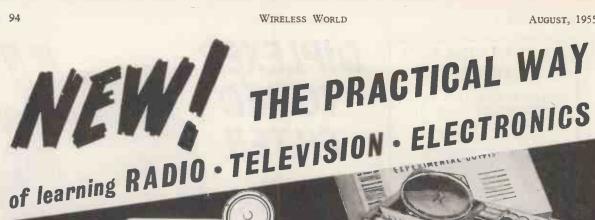
SINGLE INPUT

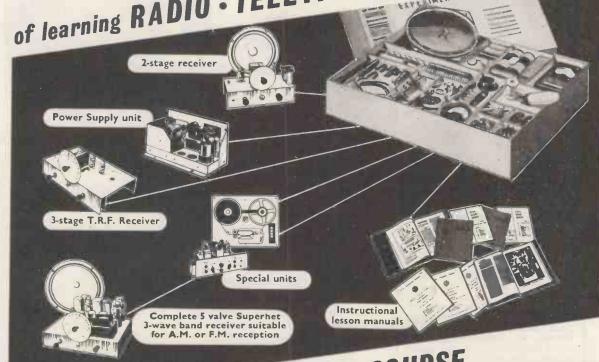
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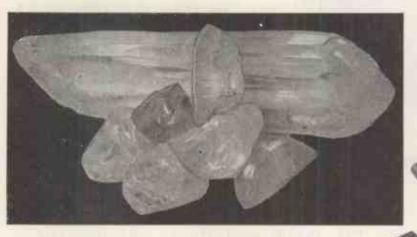
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AUGUST/55

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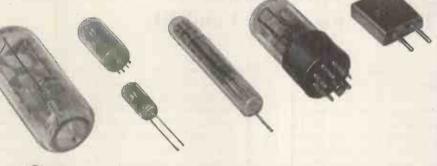
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The experience gained in manufacturing quartz crystals to the stringent requirements of our own

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ANNOUNCEMENT

FM RECEIVER ALIGNMENT GENERATOR MODEL 1324

This Alignment Generator will be available next month to provide the Service Engineer with a compact test set with which all essential alignment procedures on FM Broadcast Receivers may be undertaken.

Accurate trimming for correct overall and IF response curves is easily carried out and facilities will be provided for discriminator alignment and checks on its sensitivity and distortion.

COSSOR Model 1322

Telecheck and Marker Generator for Bands I and III

Model 1322—used in conjunction with a cathode ray oscillograph—provides equipment for the display, measurement and correct adjustment of RF and IF response curves of television receivers. This entirely new instrument comprises a swept oscillator covering the Television BANDS I and III (5-75 Mc/s. and 155-225 Mc/s.) and a frequency marker oscillator so that precise calibration of the oscillograph display may be made; accuracy of the frequency of the marker pips being verified by reference to an internal crystal. The

alignment oscillator is set to the video carrier to which the receiver is tuned and the sweep (either 1 Mc/s. or 10 Mc/s.) is automatically derived from the time base voltage of the display oscillograph. The response of the "strip" under test to the frequency band applied is then presented on the screen of the cathode ray tube. The RF output of Model 1322 is available at 80 ohms and is adjustable from a maximum of 40 millivolts to a minimum of 25 microvolts through a coarse and fine attenuator.

TELECHECK CONVERTER FOR BAND III Model 1321

This adaptor provides owners of Model 1320 "Telecheck" with an extension of the frequency range of the original instrument into the BAND III television channel. Thus, alignment procedures adopted for BAND I RF/IF "strips" are available also for BAND III receivers. A selection of the desired BAND is made by means of a switch. Pattern generator facilities for picture time base linearity checks have been retained. Model 1321 Adaptor is designed for permanent attachment to the standard "Telecheck" providing a neat, light and compact unit. Mounting is effected by four screws and the inter-connecting wiring is carried in a single insulating sleeve.

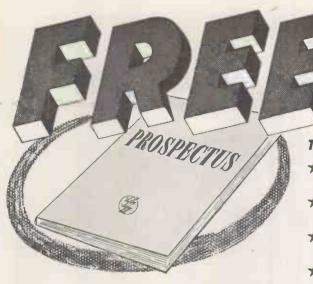


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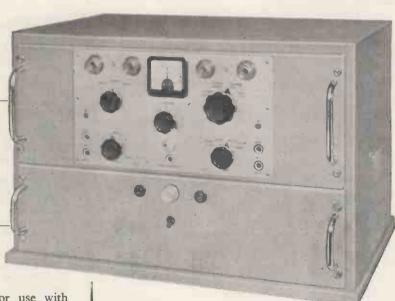
AUGUST

10382

Marconi Double Diversity Telegraph Recording Unit

TYPE HU.12

Shown here is the HU.12, in the bench mounting cabinet where it fits on top of the power unit. It can also be built into a standard rack which houses the associated receivers.



The HU.12 unit is designed for use with two suitable receivers of frequency shift or on-off telegraph transmissions in double diversity. The combined AF outputs of the receivers are converted into double current signals to operate an undulator, relay or other 'space-mark' type of equipment requiring a current up to 30 mA. No receiver modification is necessary except, perhaps, adjustment of the BFO frequency.

Satisfactory recording is possible should the first oscillator drift up to \pm 800 c/s.

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- Electronic path diversity selector for FSK will function when difference between paths exceeds 4 db.
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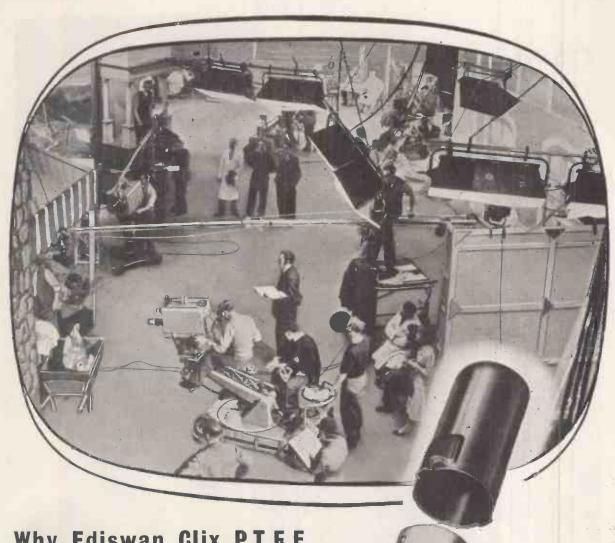
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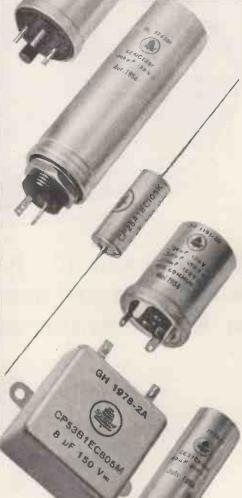
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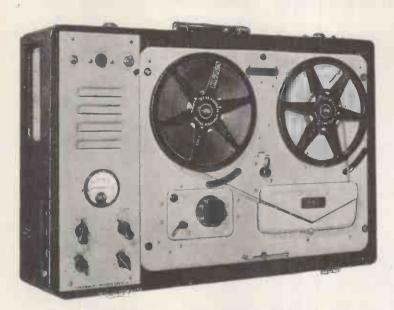
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1622

VORTEXION

TAPE RECORDER



The amplifier, speaker and case, with detachable lid, measures $8\frac{1}{4}$ in. \times $22\frac{1}{2}$ in. \times $15\frac{3}{4}$ in. and weighs 30 lb.

 \bigstar The total hum and noise at $7\frac{1}{2}$ inches persecond 50-12,000 c.p.s. unweighted is better than 50 dbs.

★ The meter fitted for reading signal level will also read bias voltage to enable a level response to be obtained under all circumstances. A control is provided for bias adjustment to compensate low mains or ageing valves.

★ A lower bias lifts the treble response and increases distortion. A high bias attenuates the treble and reduces distortion. The normal setting is inscribed for each instrument.

The distortion of the recording amplifier under recording conditions is too low to be accurately measured and is negligible.

★ A heavy mu-metal shielded microphone transformer is built in for 15-30 ohms balanced and screened line, and requires only 7 micro-volts approximately to fully load. This is equivalent to 20ft. from a ribbon microphone and the cable may be extended 440 yds. without appreciable loss.

The .5 megohm input is fully loaded by 18 millivolts and is suitable for crystal P.U.s, microphone or radio inputs.

A power plug is provided for a radio feeder unit, etc. Variable bass and treble controls are fitted for control of the play back signal.

★ The power output is 3.5 watts heavily damped by negative feedback and an oval internal speaker is built in for monitoring purposes.

purposes.

The play back amplifier may be used as a microphone or gramophone amplifier separately or whilst recording is being made.

The unit may be left running on record or play back, even with 1,750ft. reels, with the lid closed.

POWER SUPPLY UNIT to work from 12 volt Battery with an output of 230 v., 120 watts, 50 cycles within 1%. Suppressed for use with Tape Recorder. PRICE £18. 0 0.

We supply and recommend the Jason F.M. Feeder Unit. PRICE £15 17 0, including Purchase Tax.

FOUR CHANNEL ELECTRONIC MIXER

is almost essential for the professional or semiprofessional where a number of different items have to be mixed on one tape recording.

It is recommended by a number of tape recorder manufacturers for this purpose.

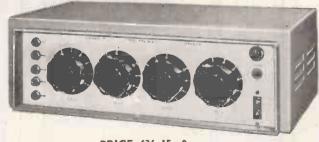
Any normal input impedance can be supplied to order, balanced or unbalanced, the standard being

15-30 ohms balanced. The normal output is 0.5 volt on 20,000 ohms or less, but 600 ohms is available as an alternative.

The steel stove enamelled case is polished and fitted with an engraved white panel suitable for making temporary pencil notes.

An internal screened power pack and selenium rectifier feed the five low noise non-microphonic

Used in many hundreds of large public address installations and recording studios throughout the world.



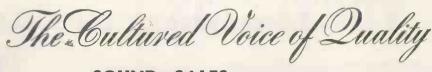
PRICE £36 15 0.

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VORTEXION LIMITED, 257-263. The Broadway, Wimbledon, London, S,W.19

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Technological Certificate in Telecommunication Engineering. At least 18 E.M.I. Scholarships are offered for the 1955 course which commences October 4th.

Dept. 127AC, 10 Pembridge Square, London, W.2.

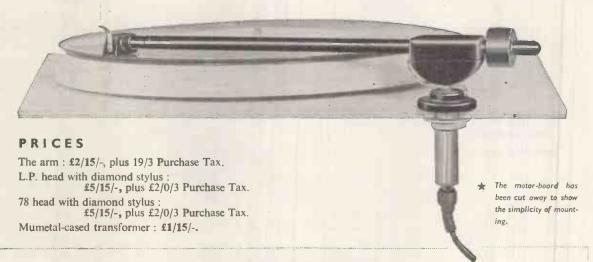
Telephone: BAYswater 5131/2

The College associated with a world-wide electronics industry, including "His Master's Voice," Marconiphone, Columbia, etc.

LEAK

NAMIC PICKUP

This pickup will very probably earn recognition as the best in the world. It is a definite advance on the pickup introduced five years ago, which is used in many recording studios for dubbing and playback.



SPECIFICATION

* THE ARM

This is of advanced design having very low inertia. Friction is kept to a minimum by using a single pivot bearing. The arm is counter-weighted and has provision for plug-in interchangeable heads. An arm-rest is

* GENERATING SYSTEM

Dynamic (moving-coil). Coll Impedance approximately 6 ohms, 1,000 c/s. No magnetic material is embodled in the moving parts, and the pickup is free from the inherent distortion of moving iron (magnetic variable reluctance) types. These distortions are also inherent in those dynamic pickups in which the moving coll is wound on a magnetic core.

* STYLUS

Material: Diamond, guaranteed unconditionally not to chip or break. Stylus sizes: L.P. 0.001 in. radius + nothing -0.0001 in. 78, 0.0025 in. radius ± 0.0001 in.

* PLAYING WEIGHTS

Between 2 and 3 grammes for L.P. Between 5 and 6 grammes for 78. Automatically adjusted by the weight of the head.

* RECORD AND STYLUS WEAR

These are lower than on any pickup of which we have cognisance. Diamond has a playing life of approx. 100 times longer than sapphire, and because it will take a higher polish than any other material it therefore causes less record wear.

The shielded step-up transformer delivers an output of 11 mV for each cm/sec. r.m.s. recorded velocity. This means that an amplifier with a sensitivity of 40 mV at 1,000 c/s will be easily loaded by the pickup from commercial records.

* FREQUENCY RESPONSE

Total variation ± 1 db 20,000 c/s to 40 c/s with the LP head, including transformer (recorded velocity 1,2 cms/sec. r.m.s. above turnover).

Low frequency resonance:
20 c/s ± 5 c/s with our very lightweight arm.

High frequency resonance:
0.001 in. radius Vynil, 21,000 c/s ± 2,000 c/s.
0.0025 in. radius on shellac, above 27,000 c/s.

The frequency response does not change with temperature.

It is not possible to specify this important ratio without stipulating the strength of the interfering fields. These fields will, of course, vary according to the installation. However, for the purpose of comparison measurements have been taken under working conditions, i.e. with various pickups mounted normally within inches of the electric turntable motor and within two feet of a power transformer in an amplifier. The results show that the Leak Dynamic Pickup has a lower hum content than any variable reluctance (moving-iron, magnetic) pickup and a very much lower hum content than a single turn moving coil (i.e. "ribbon") pickup. This confirms what would be expected from theoretical considerations.

+ DIMENSIONS

From the centre of the fixing stem to the front of the pickup head, 9½ in. From the centre of the fixing stem to the rear of the arm, 2 in. The height of the pickup is adjustable and it can be used with any turntable.

A template of original Leak design is supplied, enabling the pickup to be accurately located on the turntable mounting board. There is a single fixing hole and the stem contains a miniature socket which accepts the plug leading to the transformer (see illustration).

The transformer has a step-up ratio of 1.80 and is heavily shielded in mu-metal. The primary lead is terminated in a plug and a shielded secondary lead is supplied.

* Write for illustrated leaflet 'W'.

Sole distributors for H. J. LEAK & CO. LTD.

ELECTROSTATIC LOUDSPEAKERS

Reprints of the article by H. J. Leak, reviewing the latest advances in Electrostatic Loudspeakers, can be obtained from us on request, free of charge.

AUDIO TRANSDUCERS LTD., PICCADILLY HOUSE, 33-37 REGENT STREET, LONDON, S.W.I Telegrams: HIFI, PHONE, LONDON Cables: HIFI, LONDON

Telephone: REGent 5659

-

POTTED MAINS TRANSFORMERS

These are of really apperior construction fitted in cast metal cases and compound filled. Terminals come to ebonite baseboard. All are upright mounting and have 220/230 normal 50 cycle mains input and fully screened primary.

Type 5FI. 265-0-265 at 300 mA.: 6.3 v. at 7 amp; 4.4 v. at 2.5 amp; Price 35/plus 3/6 carriage.

Type 5F2. 365-0-365 at 150 mA.; 4 v. at 2.5 a; 6.9 v. at 4.2 a. Price 32/6, carriage and packing 3/6.

2.3 a.; 6.9 v. at 4.2 a. Frice 32/6, carriage and packing 3/6.

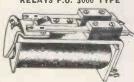
Type 5F3. 1540 v. 2 v. at 2 a.; 4 v. at 1 a.; This is an ideal transformer for televisors and scopes using V.C.R., 97, etc. Price 25/-, carriage 2/6.

POTTED CHOKES

These chokes are in similar type cases and therefore match the above transformers. Type 574, 5 ft. at 300 mA. Price 10/-carriage and packing 2/6.

Type 575, 10 H. at 150 mA. Price 12/6. Post and packing 2/6.

RELAYS P.O. 3000 TYPE



Ref. 5A1. 2,000 ohm, slow close collplat. contacts, one break, two make. Price 12/6 each.
Ref. 5A2. 2,000 ohm. standard coil, Ref. 5A1.

Ref. 5Å2. 2,000 ohm. standard coil, plat, contacts, change over make before break, two make, 1 break. Price 15/-, Ref. 5Å3. 200 ohms. standard coil plat, contacts, two make. Price 7/6 each. Ref. 5Å4. 10 ohm. standard coil, one pair plat, contacts, also mounted but not operated by the relay, are thermal changeover contacts, make before break. Price 8 6 each.

WELD TYPE WIRE JOINTER

WELD TYPE WIRE
This jointer melts the
wires and causes the
metal of each to run
together, thus making a
strong and permanent
weld. It obviously is
not intended to replace
the soldering iron but
nevertheless is ideal
for making joints that
have, for instance, to
withstand heat, vibration, chemical action,
etc.
In many cases also this
method is faster than
soldering and there can
be a considerable saving
of current. Price 9/6.
Or complete with enclosed mains transformer,
29/6.





AUTO TRANS. FORMERS

For working American equipment'off our nain etc., etc., Input tapped 200-240 v. Output 115 v. In addition to those listed be-

those listed be special this month 150/200 wat totally enclosed in metal box with input and output leaks. Price 47/6, plus 2/- post and packing.

Totally enclosed and screened.

Torally effert	sed and	i screened.	
		Price	Carr.
50 watt		: £1/2/6	1/6
100 watt		£1/16/-	1/6
			2/-
250 watt			2/6
			2/6
Unsereened			
1 KVA (1,000	w.)	£6/10/-	5/-
1.5 KVA (1,5			5/-
2 KVA (2,000	w.)	£10/17/6	7/6
3 KVA (3,000			10/-
5 KVA (5,000	w.)	£19/5/-	12/6

VARIABLE RESISTORS Heavy Duty Type

	, - Jp		365
Ohms	Amps.	Price	g
* .5	3040	35/-	(4)
*1	20-25	35/-	- 12
1.2	10-15	15/-	E
3	710	15/-	E
11	35	22/-	. [6
*50	21-4	45/-	E
*100	23	45/-	dia.
"These ar	re screw adju	t types	- 6

PAGE OF SPECIAL EQUIPMENT UNITS FOR ROTATING HEAVY AERIAL



We have brand new, still in original unopened packing cases as shipped from America two items of equipment which form part of the radar system RC84. These two units work tagether to form a Tower rotating device, with remote control.

Item 1, known as Tower 24A, is in fact the geared driving motor which rotates the mast. This is quite a heavy construction and would rotate a heavy scanner, reflector, Beam array, set., etc.

rotate a heavy scanner, reflector, Beam array, etc., etc. liem 2, known as Indicator 1-221-A is the remote controller which enables the azimuth position of Tower 24A to be controlled from a remote point. Conversely, it enables the azimuth position of the tower to be known at any time. Both the Tower and the Indicator all tis these that provide the impulses which cause the aerial to rotate backwards or forwards. The equipment intended for 117 volt A.C. mains but will operate from our mains if connected through step down transformer of 1 K.W. rating.

Prices 1-221-A £25 plus carriage. TR24A £35 plus carriage.

Special discount of £5 for cash with order or C.O.D. if both units purchased together.

R.F. HEATERS CONSTRUCTOR'S KIT

THE ELPREQ R.F. HEATER
The Elpreq R.F. Heater has been planned to
fill the need in Industry for a reasonably
priced unit to be used in the works or for

and the need a industry for a reasonably priced unit to be used in the works or for development. The header is supplied in kit form, mainly that make not love but also as it is thought that make not love but also as it is thought that make not love but also as it is thought that make not love in the love of the love to the production line.

As it is not nessible to have one frequency as the love of the love one frequency in the love of the love of the love one frequency in the love of the love one frequency.

production line.

As it is not possible to have one frequency
which is equally efficient for both dielectric
and inductive heating a frequency efficient
for dielectric work has been chosen. It
being felt that this fills the greater need.

THE POWER PACK

The Power Pack used is the "Elpreq Variable 500" which is fully described in another section, this gives ten variations of power to a maximum of 500 mA. at 1,000 V.—

continuous rating.

THE R.F. UNIT

Two carbon anode, high power triodes working into a push-pull circuit act as R.F. generators. The R.F. output to the "work" is taken from the tank coll.

Two meters are provided. The one in the main H.T. line shows the total milliamps being drawn by the R.F. unit. The other in the R.F. output stage indicates the R.F. output stage indicates the R.F.

oring trawn by one for control of the control of th

Connection to the work is through two substantial phase terminals prought out to the front panel.

Eige is approximately 16[4]in. × 13in. × 14in. and weight is approximately 25 lb. Price of all components including metal chassis to make power pack and R.F. units is £40 Or wired up ready to work £55.

All prices are ex our Eastbourne works and terms are cash with order.

RACKS AND RACKING EQUIPMENT

ALL EY-MINISTRY EQUIPMENT

STANDARD RACK

6ft. high and 19in. wide, heavy steel construc-tion. Holes drilled and tapped at the standardized spacings. Price 24/15/-plus carriage.

ENGLOSED RACK As above but rectangular and with sheet metal enclosed sides (vented), fitted handle and closing bars. Price £7/15/- or £6/15/-depending on condition, plus carriage.

MOUNTING PLATES

to fit above racks. Heavy in. steel plates (drilled at standard intervals and 19in.

centres) with chassis mounting brackets.

Ref. 5A5—19×16 front plate with chassis brackets, 17/6.

Ref. 5A6—19×12 front plate with chassis

brackets, 16/6.
Ref. 5A7—19×10; front plate with chassis brackets but drilled for meters and other

SAFETY SWITCH

When fitted this switch will cut off the mains as rack door is opened. 5/6.

CHARGING SWITCHBOARD

CHARGING SWITCHBOARD

Feed this Switchboard through a Mains Transformer and Rectifier giving 24 voit D.C. up to 50 amps, and you have an excellent multi-circuit charger for simultaneously charging several batteries at different currents. This is an ex-Gevernm. The charger are developed rated at 550 watts 18 volta fitted into steel cases with doors. It contains three reverse current relays, one voltmeter, one main ammeter, two secondary ammeters and three variable resistors for controlling circuits. These are brand new, in original cases Price £41/0-1, carriage 10/-.

We can supply a 12 volt, 50 amp. Mains Transformer at £4/5/- plus 5/- carriage.



HIGH POWER TRANSFORMERS



For R.F. Heaters, transmitters, etc., etc. These are open wound type for maximum cooling and have the normal 200-250 primary fully screened.

Type 578. 1,000 v. at 1 amp., e.g., 5. K.V.A. Price £810,--, carriage and packing 5/-.

Type 577A. 2,200 v. at 1 amp., e.g., 2 K.V.A. Price £15, carriage and packing 7/6.

7.6. The £15, carriage and packing 7/6.

Type 5M1. 1000-1000 v. at 1.5 amps. e.g., 1½ K.V.A. Price £12/10/-, carriage and packing 7/6.

Type 5M2. 1000-0-1000 v. at 500 mA. and 4 v. at 4 a. Price £7/10/-, carriage and packing 6/6.

Type 5M3. 375-0-375 v. at 250 mA. and 4 v. at 4 a. Price 37/6, carriage and packing 3/6.

Type 5J1. 500-0-500 v. at 500 mA., 6.3 v at 6 a., 45/-, carriage and packing 3/6.

POWER FILAMENT TRANSFORMERS

Type 5M4. 4 v. at 4 a. 2-0-2 v. at 10 a Price 18/6, carriage and packing 3/6. Type 5M5. 3.15-0-3.15 at 10 a., 4-0-4 at 10 a. 4-0-4 at 2-a. 4 at 4 a., 2-5-0-2.5 at 3 a. Price 27/6, plus carriage and packing 3/6. Type 5M6. 34 v. at 2 a. tapped 3/2 v., 30 v. and 28 v., for relays, etc., 22/6, plus 3/6 carriage and packing. Type 5J2. This has four 4 v. 10 amp. centre tapped secondaries. 35/5. Plus 3/6 packing and post.

POWER CHOKES. Open wound type and

POWER CHOKES. OPEN WOURD 1976— feet with Champs.

Type 5M7 30 Henry at 500 mA., 35/Type 5M8 20 Henry at 500 mA., 22/6
Type 5M9 15 Henry at 500 mA., 27/6
Type 5M10 10 Henry at 500 mA., 27/6
Type 5M11 25 Henry at 250 mA., 18/6
Type 5M12 3 Henry at 10 mmps., 18/6
Type 5M13 200 Henry at 5 mA., 15/-

PORCELAIN STAND OFF INSULATORS

threaded each end. Price 1/6



TRANSMITTER #131

This is a high powered transmitter for operating over the same frequency range as the Receiver 1132, i.e., 70-130 megascyles. It is a very bulky transmitter and probably contains around \$300 worth of equipment. As far as we know these have never been used but of course have been in store for a long time and therefore they will need attention before being put into operation. We offer these, less valves, £37/10/-, plus carriage.

R1112

We have a small quantity of these re-ceivers still available less valves. Their condition unfortunately is not good but they appear to be repairable, and, or course, contain a multitude of spare parts. At 30/- each they represent a real bargain. If not collecting, please include 5/- for packing and carriage.

AUTOMATIC MOTOR STARTER



For remote con-trol of D.C. motor between 1 and 3 kw., adjustment for 100v. or 230v. Unused and in first-class condition, complete with metal and wired glass cover, Price £10, carriage 5/-.



EX-ROYAL NAVY SOUND POWERED TELEPHONE

POWERED TELEPHONE
These require no batteries, and will go for long periods without attention. Complete with generacor and sounder which gives a high pitched note, easily heard above any other noise. Also fitted with an indicator lamp which in quiet situations can be used instead of the sounder, or where several telephones are used together will indicate which one is being called. Bize 7\(\frac{2}{8}\text{in.} \times 7\(\frac{2}{8}\text{in.} \times 7\frac{2}{8}\text{in.} \times 7\frac{2}{8}\text{in.} \times 7\frac{2}{8}\text{in.} \times 1\text{gin.} \times 2\text{if.}, wall mounting, designed for ahips' use but equally suitable for home, office, warehouse, factory, garage, etc. Price 57/8 each, plus 4/6 carriage.

BLOCK CONDENSERS



New and unused. .5 mfd. at 2,500 v., 3/6; 8 mfd. at 500/600 v., 5/-. 4 mfd. at 500 v. 2/6; 4 mfd. at 500 v. 2/6; 4 mfd. at 500 v. 2/6; 4 mfd. at 500 v. 5/-. 2/6; 4 mfd. 6/6

SENSITIVE ALTIMETER

These contain aneroid barometer move-ment and useful gears. Price 7/6 each. Post 1/-.



Note.—Also a few unused and working order, available at 22/6. and in good

SCRAMBLER-TELEPHONE EQUIPMENT

As used by Ministries and Forces for holding secret conversations. Works in conjunction with normal telephone

conjunction with normal telephone equipment. Items available, all new and unused, are;—Frequency Changer, Type 6AC, Ref. No. YB02700, price £5. Standard G.P.O. deek type instrument with scrambler switch, complete with lead and function box, price £2/10/-. Handr-inging generator in wooden box, 15/-. Junction bow with three multiple relays and cable strips, 35/-. Bank of three drop indicators in box, 15/-. Instruction book £1 refunded if returned within 14 days.

GREATLY REDUCED CATHODE RAY TUBES

VCR97. Brand new and unused, "cut-off type," ideal for 'scope, etc. Price 12/6. Carriage and insurance

5)- extriage and insurance 5)- extra. VOR517. 6\(\frac{1}{2}\)in guaranteed full picture. 29/6, pius 5/-carriage and insurance. VOR139A. 2\(\frac{1}{2}\)in, 32/6, pius 2/6 carriage, etc. VOR138. 3\(\frac{1}{2}\)in. electrostatic short persistence, suitable for T.V. and ideal for 'ecope work, 37/6 pius 3/6 carriage, etc. VOR112. 5\(\frac{1}{2}\)in. electrostatic shore persistence motatic shore and the supplementation of the supplemen

VCR112. 5in. electrostatic, persistence not known, 15/- each, plus 5/- carrlage, etc.
OV1140, CV1500, CV1546. All 12in. magnetic, long persistance, 24/10/-, plus 10/- carriage.

10-CORE CABLE

10 flexible copper conductors well in-sulated suitable for mains work. Covered Overall with hard rubber, 1/6 per yard

MORE SPECIAL EQUIPMENT HEAVY DUTY POWER PACKS

500 WATT 1,000 v. (ADJUSTABLE) The conventional circuitry is employed throughout and all components are amply proportioned to permit substantial overloading. A master switch controls the whole unit and whenever this is on current is supplied to the rectifier filaments,



current Is supplied to the rectifier filaments, thus keeping them always in the emissive state. The H.T. transformer is supplied from the primary of the filament transformer, connection being via an onlong switch controls the H.T. and the tapped choke is switch controls the H.T. and the tapped choke is switch controls the H.T. and the tapped choke is switch controls the H.T. and the tapped choke in variations from "low power" to "high power." The variations from "low power" to "high power." The variations from "low power" to "high power," to "high power," to "which otherwise would be a source of danger to users, at 505 millitamp (500 watts). But the proportions of the various components are such that 100 per cent. overloading can be allowed for pulse work or other intermittent operations. The size of the power pack is approximately 16 in. x 18in. x 18in., and its weight is approximately 87 lb. Price, Kit of parts £27/10/-, or made up ready £37/10/-.

500 WATT 2,000 v. (VARIABLE)
The maximum continuous rating of this is 250 milliamps at 2,000 volts. Rectification is half wave. Specification otherwise as for the variable 500/1,000 v.

1,000 WATT 2,000 v. (VARIABLE)

The continuous power rating of this is 500 milliamps at 2,000 voits. But the tapped choke and selector switch enables this to be reduced in ten steps. Weight approximately 120 lb., size 16in. x 13in. x 13in. Price £37/10/- in kit form, or made up ready to use £47/10/-.

1,000 WATT 1,000 v. (VARIABLE)

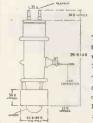
The maximum continuous rating of this is 1 amp. at 1,000 volts. Rectification is ful wave, output is variable. Weight approximately 120 lb., size 16½ln. x 13in. x 13in. rice \$37/1.0/-\$ in kit form, or \$247/1.0/-\$ made up ready to work,

FIXED MODELS

ANy of the models mentioned above can be supplied without the tapped choke and selector switch. The prices are as follows;—
Fixed 500/1,000 v. £22/10/- in kit form, or £30 made up.
Fixed 500/2,000 v. £22/10/- in kit form, or £30 made up.
Fixed 500/2,000 v. £32/10/- in kit form, or £40 made up.
Fixed 1,000/1,000 v. £32/10/- in kit form, or £40 made up.
All prices quoted are ex Works.

CEILING FAN

This model, made by Revo, incorporates a series-wound totally enclosed ball-bearing motor of robust construction and noiseless operation. The fan has a blade diameter of 36in. and is supplied with 20in. suspension tube and ceiling canopy. All finished white cellulose enamel. The voltage working is 230-250 v. D.C. Revo catalogue number D12288 Price £10/10/-



SPECIAL PURPOSE VALVES

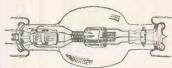
Triode Type CV 1098-this is a high-power air-cooled Specification of which is as follows; Filament voltage 8.2 v., filament current 35 amps., anode dissipation 750 watts. Maximum anode voltage 23 kV

This valve is very suitable for R.F. heating at high frequencies and two of these in push-pull under Class C conditions would have an output of approximately 2 kilowatts. Brand new, still in original shockproof packing, price

TETRODE TYPE VT31

This is a high-powerd air-cooled tetrode. Specification of which is as follows:—Heater volts 11.25, heater current 8 amp., maximum anode voltage 5 kV., anode dissipation 250 watts, size approximately 14in. long and 6in. across the builb.

bulb.
Limited quantity only at £4 each, still in original packing



WATCH THESE COLUMNS FOR DETAILS OF VARIOUS OTHER INDUSTRIAL TYPE VALVES.
ALTERNATIVELY SEND US YOUR ENQUIRIES.

HIGH CYCLE MOTOR ALTERNATOR

TYPE 1. Has a motor 230 v., 50 cycle single phase 2,800 r.p.m., coupled to a generator output 250 v., 1,728 cycles at .24 amps. Good condition, with wiring diagram, £3/10/-.pha 7/6 carriage.

TYPE 2. Has a motor 230 v. 50 cycle single phase, coupled to an alternator output 250 v. 625 cycles .24 amps. Price £3/10/-, plus 7/6 carriage. SPECIAL EQUIPMENT SALES

IMPORTANT NOTE

Owing to the bulkiness of many of the items listed on these two pages it may not be possible to keep stocks at branches, therefore please telephone confirmation that the item is actually at the branch before journeying specially to see it.

SPECIAL SALES DEFT., E.P.E. LTD., BOURNE HOUSE, GROVE ROAD, EASTBOURNE.

Phone: Eastbourne 5055



"SNIPER-SCOPE

Famous wartime "cat's eye" used or seeing in the dark. This is an infra-red image converter cell with a silver caesium that the control in the control in the cathed ray steel which lights up (like a cathed ray steel which is the control in the cathed cathed the control in the cathed cathed the cathed c



0.300 mA. moving coil 10.500 mA. moving coil 10.500 mA. moving coil 1'	0/6 0/6 0/6 7/6 7/6
0-3 amp. R.F. thermo 0-5 amp. R.F. thermo 0-5 mA. moving coil 0-20 amp. moving coil 10	7/6 7/6 7/6 8/6 8/6 8/6 2/6
	2/6

PYREX AERIAL INSULATORS

Ideal for aerial con-nections through cabin walls or through pan-els. Consists of glass dome with threaded rod and terminal ends and metal fring and metal fixing flange. Price 2/- each fixing



PLUG AND SOCKET





This brass cased plug and socket is extremely robust and ideal for P.A. or outside work. Ideal also for taking power to units as it insulates the ends of the control of th



JUMBO VALVE BASES

Ceramic 4-pin for transmitting valves. Price 3/6 each.

FLEXIBLE COUPLINGS



These are somedimes known as bellows couplings extend as well as bend. They are ideal for joining shafts which are out of alignment and for slug tuning controls where the core has to come in and out. Price 1/9 each.



30 AMP ROTARY SWITCH

Single pole ON/OFF, a very robust switch. made by one of our most famous firms. Will give life-time of service. Price complete with pointer knob, 4/9.

41/6

42/6

561-52/6 53/6

67/-

69/-

83/6

134/-

145/-178/6

94/6

AERIALS

BAND III AERIALS These aerials have quick fitting ele-ments, all alloy tube construction and polythene low-loss insulators.



DOWNLEADS

BAND III DOWN-LEAD. Patented five cell construction ensures maximum air polythene ratio around the con-ductor, high perductor, high per-formance, 9d. per

THIS MONTH'S SPECIAL

The first of a series of highly efficient low priced aerials to be made by Cleveland Electric. This is an array for Band III technically, known as a folded "v" it is highly efficient, and directional and suitable for indoor or outdoor use, even for fitting on the window frame or coping, or existing T.V. mast, it employs polythene low loss insulators and non-corrodable elements.

Heavy demand likely, so please order at once. Price 17/6, post and packing 1/6.

and packing 1/6.

PRE-ASSEMBLED WITH fin. DIA. RODS

DESCRIPTION

3-element array with swanneck mast with 'U' bolt clamp for fitting to exiting masts from #in. to 2 ln. dia...

3-element array with cranked mast and wall mounting bracket. List Price o-element array with cranked mast and chimney lashing equipment
5-element array with swanneck mast and 'U' bottclamp for fitting to existing mast from in te 2in. dia.
5-element array with cranked mast and wall mounting
bracket
5-element array with cranked. bracket
Selement array with cranked mast and chimney lashing
equipment
Selement array with swanneck mast and 'U' boit
clamp for fitting to \$\frac{1}{2}\text{in}\$. to \$2\text{in}\$. dia. mast.

Selement array with cranked mast and chimney lashing

equipment
8-element array with lin. mast cap, 10ft. mast and
heavy duty single chimney lashing equipment
10-element array with cranked mast and chimney

WHAT IS IT?

It is the indicator that you would make to check that the "Elpreq Band III Signal Generator" is working properly. When the loop is brought up to the output circuit the 'amp lights brightly.



THE "ELPREQ" Band III SIGNAL GENERATOR is most useful. It:-

- THE "ELPREQ" Band III SIGNAL GENERALUK IS MOST USETM. III:—
 1. Will provide the signal for tuning to any Band III station.
 2. Can be used as a grid-dip meter for checking the frequency of Band III T.V. aerisls, Colls, etc.
 3. Can be made to give a pattern on T.V. Receiver screen.
 4. Can be accurately calibrated with included equipment.
 All the parts including valves, tuning condenser and metal chassis are available as a Kit at 25/- post free. Constructional data free with Kit or available separately price 2/6.



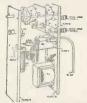
BE READY FOR THE NEW COMMERCIAL T.V.

Daily Test Transmissions are already taking place

Our convertor which fixes to the side or back of your T.V. will give you the new station or the old by the flick of a switch. You do nothing to your existing set; just plug in mains and aerial leads. Suitable for any T.V. Price 25(7.0)-, or 30/- deposit and six payments of £1.

RUILD YOUR OWN CONVERTER

The Converter has given very satisfactory results from the experimental Beulah Hill station. It uses 2 valves, is not at all difficult to make and can be lined up with the simple 25/- instrument described above. Price for all the components including constructional data is £3/10/- or £4/10/- with mains equipment. Price includes stove enamelled case, prepared metal chassis, and all parts to make up convertor as illustrated-data available separately price 2/6.





MADE FOR THE **JOB**

This ex-W.D. 10-valve superhet was designed to receive 200 megacycles transmissions so it will require virtually no conversion to receive the commercial T.V.

These contain 6 valves type SP61, and one each RL7. BL16 and EA50. Six IF transformers 12 Mc/s. band and hundreds of other useful components. Price 59/6. plus carriage and packing 7/6 These receivers are unused and perfect.

THE ELPREQ F.M. UNIT

In the ELPREQ F.M. Therefore valves and two crystals are used. The last valve arts as a limiter so reducing the necessity of exact tuning and at the same time improving interference interference with the ratio detector to mode improving interference the countries of with used in the ratio detector to mode heater-cathode hum so often encountered with used in the ratio detector. Stability is oxtremely good and tuning most simple. The tuner draws the low over supplied from the set or amplifier, its valve heaters are not connected to earth. With only a simple indoor aerial made by parting the ends of ordinary flexible cable this tuner works very well at Eastbourne (over 60 miles from London) and we await reports from even greater distances. Cost of all parts including valves, prepared metal chassis, scale, slow motion drive, pointer, tuning knob, in fact everything needed to make the complete unit suitable for mounting through the side or back of an existing reasonably sized radio, or into a separate cabinet, is £6/12/6, data is included free with the parts or is available separately price 2/-.

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3-speed record player with pick-up using the famous Acos "Hi G" turnover crystal —motor also by very famous maker. All on unit board ready for installation. A wonderful bargain at £5/10/-, plus 5/carriage.



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Veneered and Polished-Perfect. New and unused.

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50 assorted \(\frac{1}{2} \) and \(\frac{1}{2} \) wattresistors. Ranging between 10 ohms and 10 meg. ohms. (Our Selection). Price 5/- pkt. 50 at 1 watt. 7/6.

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Five valve superhet chassis—size 15 5 × 2in, with three waveband glass scipulley, driving head, etc.

Also TABLE CABINET Veneered and polished—takes the above chassis, 39/6. Post 2/6.

NOVELTY RADIO

Complete tunable M/L Radio with room for 3in. speaker in base. Needs only valves, speaker and batteries, 29/6, plus, 2/6 postage, etc.



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a Chassis size
approx. 9\frac{1}{2} \times 7\frac{1}{2} \times 8\frac{1}{2}. First-class components. A.C.
mains operation. Three wave (medium and two shorts). Complete with five valves, ready to work. Special ce this

cash-with-order price this month. \$5/19/6, carriage and insurance 7/6.



BAKELITE CABINET

Two tone with built-in handle. Note.—All have slight imperfections but these are hardly noticeable and will not affect the soundness of the set.



Post 9d.

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INSTRUMENT CASE Veneered and polished-undrilled.

14′6 2/6.

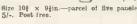
CONSTRUCTOR'S PARCEL

Bakelite cabinet, complete with d metal chassis and back, and plans T.R.F. dial



UNBREAKABLE GLASS PANELS







CORNER CONSOLE

A massive cabinet but being corner fitted is not out of place even in a modern small living room. Overall dimensions of this cabinet are 47in. wide × 31in. (deep to corner) × 50in. high. Made to house 15in. Televisor, Radio Unit, Amplifier, Tape Deck, etc. Originally £18. Our Price-£10 plus 30/- carriage.

36

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THIS MONTH'S SNIP

witch, and scale.

These are ex-equipment, fully reconditioned tested and guaranteed. Price coaly 14/6 (less than the value of coil pack only), complete with circuit diagrapost and packing 1/8 extra.

SELECTIONS FROM MISCELLANEOUS STOCKS

TRANSORMER PRIMARIES
These are transformers with wound, primaries tapped 200, 220 and 240 but with no secondaries. There is ample window space, however, for the hand winding of secondary to suit your own requirements. Number of turns per volt required, depends upon the overall size of the stack, for example, our 100 watt stack requires approximately 2½ turns per volt. The size required depends upon amps. taken out and the voltage eg. if 100 amps. at 10 volts is required then a 100-watt primary is needed. 50 Amps. at 5 volts requires 250 watt, etc., etc. Transformers are complete with clamps and feet. TRANSORMER PRIMARIES

clamps and	feet.	
Watts	Price	Post & Pkg.
80	7/6	1/6
100	10/-	2/-
120	12/6	2/6
250	12/6	2/6
500	25/-	4/6
1,000	40/-	6/6
2,000	30/-	8/6

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Stud Switch for building up variable resistors as illustrated or similar devices, complete with handle, metal side pieces, and spacer bolts. Price 6/6, post and packing 1/-.

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SELECTIONS FROM OUR RANGE OF CABINETS



This cabinet is undoubtedly a beautiful piece of furniture. It is elegantly veneered externally in figured walnut, internally in figured walnut, internally in white sycamore. The radio section is raised to convenient level but is not drilled or cut. The lower deck acts as the motor board, again is uncut, it measures 16×14im. and has a clearance of 5in. from the lid. There is a compartment for the storage of recordings.

Overall dimensions of this essentially modern cabinet are 3ft. wide, 2ft. 8in. high, and 1ft. 4½in. deep.

Price £15/15/-, carriage etc., 12/6.

THE CONTEMPORARY Also in the modern trend is this

wery stylish contemporary con-sole. Veneered in oak with con-trasting mouldings, and is ideal for use with modern furniture or with other contemporary fittings or furnishings. The radio and or furnishings. The radio and motor board is uncut and its size, 30×15½ in., provides ample room for all equipment. Price £8/15/-, carriage etc., 12/6.



OFFERED AT APPROX. HALF COST TO MAKE

An impressive costly looking cabinet originally designed for T.V. but simple originally designed for T.V. but simple modification makes the cabinet suitable for radiogram, amplifier, tape recorder, or reflex speaker—size 23in. wide, 22in. deep and 37½ in. high. Limited quanity left at £8/15/- each, which is approximately half of their manufacturing cost. Also slightly damaged but repairable. Price from £6. Carriage cost. A pairable. 12/6.

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5-Valve 3-wave band superhet covering long, medium and short wave.
Osram miniature valves are employed and low loss from cored coils account for an excellent signal-to-noise ratio. Full A.V.O. is applied to both frequency changer and I.F. stages.

stages.

The output stage utilises variable negative feedback. A gram, position is provided and reproduction of record-

sole negative reconsist. A gram, position is provided and reproduction of reconsis particularly good. Chassis size is $12 \times 7 \times 71$ m.—scale sizes is $10 \frac{1}{2} \times 4\frac{1}{2}$ m. This reciver has been tested in particularly difficult areas and its stability and noise rejection have produced exceptional results. Price $211/10/10 \cdot \sigma \in 23$ deposit—carriage, etc., 7/6 Similar model but with ferrite rod aerials 212/10/10. Hire purchase deposit 24.



This is the Model F3PP. Developed especially to meet the Increasing demand for high fidelity equipment. Particularly suitable for replacement in a radiogram. This is a 7-valve 3 wave band superhet with push-pull output, incorporating separate bass and treble controls thereby ensuring a maximum control of fidelity, volume and tone. Wave band coverage 16-50, 190-550, 900-200 metres. Valve line X79, 6Ba6, ECSS, GZ30 and two 6AQ5. This chasse is suitable for use on AcC. mains from 100-110 v. and 200-250 v. Price 17 guiness or £4 deposit carriage and insurance 7/6.



ANOTHER CLEVELAND CHASSIS-"THE TREMENDO"

ANOTHER CLEVELAND CHASSIS—"THE TREMENDO".

The Cleveland Organtone is good, but this one is really superb. It has a 7-valve circuit with 6 watts output, litted with independent bass and treble controls. It is really an efficient R.F. circuit coupled to a high-fidelity amplifier. The chassis size is the same as the Organtone, namely 12×7×7 with the 10½×1½ multi-coloured scale, and it is built to the same exacting specification as the Organtone. Price £15/10/-, carriage and packing 7/8. H.P. terms if required. Ditto but with ferrite rod aerial coil. £16/10/-.

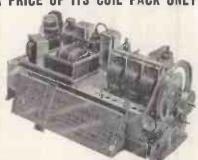
ARMSTRONG THE

Among high class radio chassis, the most famous, and their new model FC48 certainly lives up to tradition. It is virtually a 10-valve circuit, for among its eight valves two double triodes are employed. Special features of this chassis are (a) 8 watts output in a push-pull circuit with ample negative feed-back to ensure the highest fidellity: (b) provision for using F.M., e.S., power brought out to sockets and indicator on dial: (c) independent base and treble controls with visual indication of setting: (d) four wave bands covering 16-51, 50-120, 190-550, and 1,000-2,000 metres. The size of this chassis is 124-x9x94in. Price £23118/-, plus 7/6 carriage and insurance.



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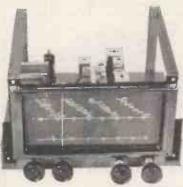


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Model "A," FM tuner. A popular and small unit, with good sensitivity. These are in use from Bognor to Ely, and little changed since first described by Amos and Johnstone in the "Wireless World." New "hammer" finish front plate and tuning scale carries a magic eye; this and power unit are optional.

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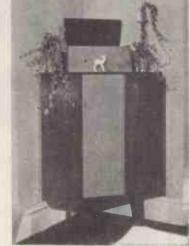
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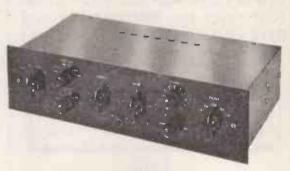
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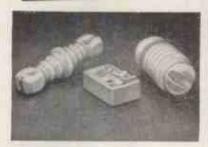
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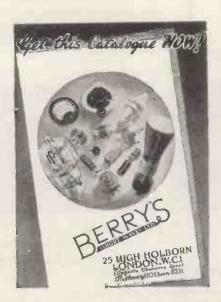
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- * Two-stage Capstan
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- * Independent Bass and Treble
- ★ Twin-track recording
- ★ High Fidelity Amplifier

can be used independently of the recorder for P.A., Radio or Record Reproduction.

The NEW SIMON Model SP/2 Tape Recorder is designed and built to top standards. Years of specialist experience in the field of sound-recording engineering and techniques have been combined to produce an entirely new equipment which is faultless in performance and appearance. The SP/2 provides superb recording and reproduction facilities. Ask your dealer to show you the new SP/2 now. . . .

The inside story . . .

Ask for a copy of the new booklet—"Affairs of Tape"
—free, of course. Brings you up to date on Tape
Recording—gives you the inside story of the SP/2.

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46/50 GEORGE STREET, LONDON, W.I. WELbeck 2371 (5 lines)

="You can rely on us"

Stockists of all Radio and Electronic components for manufacturers, laboratories, Educational authorities, and the amateur.

Stockists of Bulgin, Denco, Hunt, T.C.C., Belling and Lee, W.B., Whiteley, Wharfedale, etc., etc.

Mullard 510 amplifier and G.E.C. 912 amplifier—all parts stocked and available on H.P.

Taylor and Avo instruments from stock. (H.P. terms available.)

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82, SOUTH EALING ROAD, LONDON, W.5.

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Next to South Ealing Tube (TURN LEFT) 9 to 6 p.m. Wed. I o'clock.

RECEIVER CHASSIS

Modernise your old Radiogram

RECORD PLAYERS

COMPLETE RADIOGRAM EQUIPMENT-QUALITY AT LOW COST

STERN'S DESIGN FOR HOME CONSTRUCTORS This 3-SPEED AUTOCHANGER is by

The "SUPER-SIX"

A compact and highly efficient superhet Radio-Radiogram chassis of outstanding quality.

BUILDE YOU CAN IT FOR £10/7/6

Including the OCTAL VALVE LINE-UP (£12/7/6 with the miniature valves).

(£12/7/6 with the miniature valves).

Incorporating the new B.V.A. Miniature Valve Lineup. This receiver is designed to the very latest specification and provision is male to incorporate either the standard
form of the provision of the property of the production of both Radio reception and
form attention has been paid to the quality of the reproduction of both Radio reception and
feecord playings, and excellent clarity of speech and music is obtained.

A few brief details.

Covers 3 wavebands 18-50 metres, 190-550 and 800-2,000 metres.

Employs 6 valves having PUBH-PULL for 5-6 watts output.

Incorporates delayed A.V.C. on all wavebands and pre-selective feedback.

A 4 positi. In Tone Control operation on both Radio and Gram.

Has independent mains supply socket for a Record Player.

Size of Assembled Chassis 12lm. × 8lm. x 8lm. Dial aperture 8lm. x 4lin.

For operation on A.C. mains 200-250 volts 50 cycles.

THE INSTRUCTION and ASSEMBLY MANUAL is available for 1/6. It contains very
detailed practical drawings and circuit diagrams and a complete Component Price List.

SUPERHET RECEIVER For only

(Carr. & ins. 7/6 extra)

H.P.—23/4/6 Dep. 12 mths. at 18/4.

These receivers Models AW3-7 are made by a well-known set of manufacturars and incorporate the latest Osram Valve Interuporate Osram Valve Interuporate the latest Osram Valve Interuporate Osram Valve Interuporate Interuporat

A BULK PURCHASE ENABLES US TO OFFER THIS "PUSH-PULL" SUPERHET RECEIVER £12/19/6



!!!THE LATEST!!! RADIO-RADIOGRAM CHASSIS

Model F3PP. A 7-valve 3-waveband Superhet Chassis with a Push-Pull Stage. This Chassis has been designed with particular regard to the quality of reproduction. It incorporates SEPARATE BASS and TREBLE CONTROLS, thereby ensuring the utmost flexibility of Tone on both Radio and Gram.

WE RECOMMEND THE W.B. "STENTORIAN" P.M. SPEAKERS They have the NEW CAMBRIC CONE and a matching device for 3 ohm-7.5 and 15 ohm outputs.

(a) 5in. Model HF510 (b) 6in. (d) 9in. Model HF912 (e) 10in. ,, HF1012 (f) 12in. ,, HF1214 £1 19 6 £3 9 6 £3 17 6 £9 15 6 .. HF610 .. HF810 £2 12 6 £3 5 6 (c) 8in.

OTHER TYPES IN STOCK
m V/Coil 15 9 8in. P.M. 3 ohm V/Coil 18/9, 19/6 & 25/n
TV/Coil 16 6 10in. P.M. 3 ohm V/Coil 25/- & 37/n
m V/Coil 16 9 12in. P.M. 15 ohm V/Coil £5 5
All are NEW and FULLY GUARANTEED 3½in. P.M. 3 ohm V/Coil 5in. P.M. 3 ohm V/Coil 6½in. P.M. 3 ohm V/Coil 25/- & 37/6 £5 5 0

Famous Manufacturer and is offered for £9/19/6 (Plus 7/6 carr. & ins.) Normal Price £13/10/-.

Purchase Terms Dep. and 9 months at 19/-.

These units will autochange on all three speeds. 7in., 10in. and 12in.

They play MIXED 7in., 10in. and 12in. records.

They have separate sapphite for

12in. records.

They have separate sapphire for L.P. and 78 r.p.m., which are moved into position by a simple switch.

Minimum baseboard size required 14in. x 12in., with height above 5in. and height below biseboard 2iin. A bulk purchase enables us to offer these BRAND NEW UNITS at this exceptional price.



WE CAN ALSO SUPPLY

THE GARRARD 3-SPEED CHANGER MODEL R.C.80M.

(a) THE GARRARD 3-SPEED NON-AUTO MODEL "T".
(c) THE COLLARO 3-SPEED NON-AUTO MODEL R.C.54.
(d) THE COLLARO 3-SPEED NON-AUTO MODEL 3/554.
(e) THE B.S.R. 3-SPEED NON-AUTO MODEL 13/554.
(e) THE B.S.R. 3-SPEED NON-AUTO MODEL HFi00.
All of these are the very latest models. Send S.A.E. for details.

WE CAN SUPPLY FROM STOCK



The COLLARO 3-SPEED "TRANSCRIPTION" **PLAYERS**

MODEL 2010 with the NEW LGHTWEIGHT "STUDIO P" CRYSTAL PICK-UP.

PRICE £18/5/3 (plus 7/6 carriage and insurance).

H.P. TERMS: Deposit £4/11/3, 12 monthly payments of £1/5/5.

MODEL 2000 (few only) 3-speed unit only (excluding pick-up).

PRICE £13/9/6 (plus 7/6 carriage and insurance).

H,P. TERMS: Deposit £3/7/6 and 12 monthly payments of 18/10

THE NEW ARMSTRONG F.C. 48

A high quality replacement Radio or Radiogram Chassis having provision for an F.M. Feeder Unit. PRICE ASSEMBLED AND READY

£23/18/0 FOR USE

(Plus 7/6 Carr. and Ins.) H.P. Terms: 25/18/- Deposit and 12 months at £1 13/9.

OUTSTANDING FEATURES INCLUDE:-

8 Valves including 2 double Triodes.
8 Watts output from push-pull tetrodes. Heavy negative feedback is used, resulting in negligible distortion and high dampling factor.
Provision for using FM dadport to receive the present high quality transmissions from Wrotham and the new B.B.C. V.H.F. stations.
An accessible socket at rear provides the power stupply for this unit.
Independent controls give BASS and TREBLE lift and cut with unique Thermometer visual indicator.
Gram. position on a wavechange switch.
4 Wavebands Coverage 16-51, 50-120, 190-550, 1,000-2,000 metres.
Large four-colour illuminated dial.

WE HAVE THE NEW ARMSTRONG F.M. FEEDER UNIT Model F M 55

One of the best F.M. units in production. Consisting of a 5-VALVE SUPERHET DESIGN with frequency coverage 83-85 mc/s. Power supplies required 250 volts at 30 m/a and 5.3 volts at 2 amps. Price 221 (plus 7/6 carr. and insur.). H.P. TERMS; Deposit 25/5/- and 12 months of £1/8/4.

SPECIAL REDUCTIONS FOR COMPLETE EQUIPMENT

SUMMARY—Select a RECEIVER CHASSIS and we will supply it TOGETHER WITH A 3-SPEED CHANGER AND AN 8in, or 10in. P.M. SPEAKER as follows:—

THE B.S.R. MONARCH, P.M. SPEAKER and:-Cash Price Denovit Monthly £1 11 11 £1 13 4 £2 0 1 £2 9 2 THE COLLARO MODEL R.C.54, P.M. SPEAKER and :-14

GREATLY REDUCED——WE OFFER THE ACOS "MIC 22-2" CRYSTAL MICROPHONE

This is a High Fidelity Mike incorporating the "Filtercel" insert and normally retails at £4/4/-. It is complete with Table Stand. (plus 2/- carr. & insur).

LTD.

UNITS

RADIO TUNING High Fidelity A.M. & F.M. DESIGNS Reproduction

AMPLIFIERS ASSEMBLED or KITS OF PARTS

Having a front panel which is very attractively finished in deep gold, and on which the controls are clearly identified. The ideal amplifier for general home use and for small halfs, etc. Price of COMPLETE KIT Including Valves and Drilled Chassis, etc. (Plus 2/6 carr, and ins.)

We will supply it Completely Built

We will supply it Completely Built for (Plus 5/- Carr. & Ins.) \$9/10/(Plus 5/- Carr. & Ins.) \$29/10/Designed for high quality reproduction up to an output level of 10 watts. having \$9'0's in Prush-Puil and incorporating negative feedback, It is suitable for use with all types of Pick-ups and most types of microphones and the output transformer provides for use of 3 and 18 ohm speakers.

BRIEF FEATURES
Valve line-up-6Z5, 6SN7, 5Z4, with 9V6s in push-pull.
The undistorted output level of up to 10 watts is produced from an input of .25

voits. First class reproduction of Radio (where a Tuning Unit is used) and Record Playing. Separate Bass Boost and Treble Controls provide an excellent range of frequency

control. Very satisfactory results are obtained with an average type of high impedance Moving Coil or Crystal Microphone, a clear speech level of approx. 5 watts output

being obtained. Great and LT) are available for a Tuning Unit.

Owner supplies (HT and LT) are available for a Tuning Unit.

For operation on A.C. Mains 200-250 volts 60 cycles.

THE ASSEMBLY MANUAL is available for 1/s and includes detailed layouts and component Price List.

"STERN'S" HIGH QUALITY 8-10 WATT AMPLIFIER "STERN'S" 12 WATT "HIGH FIDELITY" Push-Pull AMPLIFIER

"STERN'S" 12 WATT "HIGH
A very high quality Unit attractively dinished in
deep gold with each control clearly identified on
the front panel. Comprising a Main Amplifier
Chassis and a Remote Control Pre-Amplifier-Tone
Control Unit. The remote control unit
measures only 9 × 4 × 2½in. and
contains four controls, being; BassTreble-Volume and a Radio, Gram,
Microphone Switch control.
If incorporates its own feedback circuit on the Bass
Channel. Loop negative feedback is employed on the
Main Amplifier which has a
valve line-up of 635-6N7-5U4
with two PX85s in push-pull
and 635 and 63N7 are used
in the remote control unit.

THE COMPLETE KIT IS AVAILABLE FOR (Carr. & Ina. 6/- extra.)

£14/0/0 THE COMPLETE UNIT ASSEMBLED AND READY FOR USE \$116. H.P. Terms 24/5/- Deposit, 12 Months at £1/3/11. (Carr. & Ins. 7/6 extra.)

H.P. Terms 24/5/- Deposit, 12 Months at 21/3/11. (Carr. & ins. 7/6 extra.) The measured frequency range of the amplifier with this unit shows an excellent response from 14,000 cycles down to 20 cycles, the bass and treble controls allowing independent control of gain at both ends of the frequency range from zero to a gain of 50. It can be seen, therefore, that ample correction is provided to suit any type of pick-up with any type of recording. Input voltage for maximum output is 70 mV and 6.3 volts at 2 amps. and 30 mA. H.T. is provided for tuning unit, etc. This Amplifier compares well with the Williamson and similar designs at a fraction of their cost. The complete set of assembly instructions is available for 2/*.

A COMPLETELY ASSEMBLED "HIGH FIDELITY" PUSH PULL

AMPLIFIER. Supplied Complete with THE STERN'S DUAL CHANNEL TONE CONTROL AMPLIFIER. PRE-AMPLIFIER UNIT FOR ONLY £13/13/-

(plus 7/6 Carr. & Ins.)

H.P. TERMS DEPOSIT £3/8/- and 12 monthly payments of 19/2.

We are able to offer this equipment at such an attractive price only because of a bulk purchase of PARMEKO TRANSFORMERS, CHOKES, etc.

It is daskings of the price only because of a bulk purchase of PARMEKO TRANSFORMERS, CHOKES, etc.

It is daskings of the price of the production, employing two 6Ffis in push-pull for approximately 10 watts output. A total of 7 valves are employed, the main Amplifier having 615—68N7—4wo 6F6s and 5 Volt Rectifier and the separate Control Unit, which is identical to that supplied with the 12 Watt. "HI F" Amplifier described above, has types 635 and 68N7. Loop Feedback is employed over the whole of the main Amplifier and the PARMEKO OUTPUT TRANSFORMER ensures really good reproduction. Power take-off socket is provided for an external Radio Tuning Unit. production. Power take-off socket is provided for an external Radio Tuning Unit, the POWER SUPPLY AVAILABLE being 200 to 250 Volts at 45 mA, and 6.3 Volts at 14 amps. WHEN ORDERING PLEASE STATE WHETHER FOR 3 OR 15 ohm SPEAKER.



THE DENCO F.M. FEEDER UNIT INCORPORATING AN R.F.

A 5 VALVE SUPERHET DESIGN having A D VALVE SUFERHAL DESIGN BOWN of a frequency coverage of 88 to 100 mc/s. This F.M. Receiver is designed to operate with any type of Amplifier and most Radio Receivers. It incorporates R.F.—F/Changer and two I.F. Stayes followed by a Ratio Discriminator, the valve line-up being 6AM6. — 124 H8—40 6BA6s and 6AQ5. Overall size of assembled Chassis 7In. x5jin.x

size of assembled Chassis 7ln.x5lin.x 4lin. high excluding power supply, or 7ln.x8lin.x4lin. high with power supply. The CONSTRUCTOR'S MANUAL, containing Circuit Diagram and Component Layout, etc., is available for 1/6, and WE CAN SUPPLY ALLSPECIFIED COMPONENTS including Valves and Drilled Chassis for Assembly as illustrated.

or for £7/2/6 with Dial £5/13/6 (pius 2/6 carriage and ins.)

WE WILL ALSO SUPPLY IT .

(a) Assembled and Ready for use, excluding Dial Assembly, £8/17/6.

(b) Assembled and Ready for use including Dial Assembly (as illustrated), £9/10/-. (c) Assembled and ready for use, with Dial Assembly and "Magic Eye" indicator mounted in centre of Dial. £10/10/-.

(d) We can also supply (a), (b), and (c) with and including an HT LT Power Supply for an adultional £2/17/6. The Supply Unit is also available as a separate Unit size 6\(\frac{1}{2}\) in x 3\(\frac{1}{2}\) in h \(\frac{1}{2}\) in color at 5\(\frac{1}{2}\) on an 6\(\frac{1}{2}\) or list at 5\(\frac{1}{2}\) and 6\(\frac{1}{2}\) or list at 2 amps

109 × 115 FLEET ST

LONDON. E.C.4. Phone: CENTRAL 5812-3-4



The NEW "LEAK" TL/10 AMPLI-FIER and "Point One" PRE-AMPLIFIER

This Amplifier has a maximum output of 10 watts and maintains in every respect the world renowned LEAK reputation for precision engineering; fine appearance and fastidious wiring. The Pre-Amplifier will operate from any make or type of plek-up. A continuously variable input attenuator at the rear of the Pre-amplermits the instantaneous use of crystal, program from and moving cell pick-up. moving iron and moving coll pick-ups. H.T. and L.T. supplies are available for a Radio Tuning Unit. An input attenuator is fitted. S.A.E. for descriptive leaflet.

PRICES :

(a) The COMPLETE AMPLIFIER WITH PRE-AMPLIFIER, £28/7/-, or £7/2/-

(b) The TL'10 MAIN AMPLIFIER ONLY: £17/17/-, or £4/7/- Deposit and 12 months at £1/5/4

(c) The "POINT ONE" PRE-AMPLIFIER ONLY: £10/10/-. or £2/12/6 Deposit and 12 months at 15/-.

"STERN'S" MODEL CP3G 3 WAVEBAND SUPERHET TUNING UNIT

A highly sensitive tuning unit providing for excellent reception of stations on the short wavebands (16-50 metres) medium wavebands (16-50 metres) and the long waveband (800-2,000 metres). We can supply this tuner to correctly operate with each of the Amphillers.

Valve line-up; 6886 (Frequency Changer)
68K7g (i.F. Amplither), 697g (Detector.
A.V.C. and 1st A.F. Amplither), and 524g (rectifier).

A granophone position is incorporated with the wavechange switch and the 697g valve becomes the last A.F. Amplither for the gramophone pick-up.

This tuner is normally supplied with four controls—Tuning, Volume, Tone and the Wavelength Switch (Tone and Volume operate as both Radio and Gram.)—but if your Amplither already has the Tone and Volume Controls we can omit both. When ordering please state what is required.

Overall chassis dimensions are 12in.x8in. including the full vision dial. Size 8ijn.x4jin.

For A.C. Mains only, power supply required—H.T. 250 volts 30 mA., L.T. 6.3 volts if amp.

ce. completely assembled and including built-in power supply £10/10/-. H.P. rms. Deposit £2/12/6. I2 months at 15/-,

Price completely assembled excluding Power Supply £9. Carriage and Insurance 7/8 extra. (Dial Escutcheon is 4/8 extra.)

SELECT ANY TUNING UNIT and an assembled AMPLIFIER (or a TUNING UNIT, AMPLIFIER and BECORD PLAYER) and we will quote you a REDUCED PRICE for the combined order. H.P. TERMS ALSO QUOTED. . . . As an example

we offer:
(a) STERN'S 8-10 WATT AMPLIFIER and the Model CP3G (or DENCO F.M. TUNER) all
assembled for £17/-f- (plus 10)- carriage & insurance). H.P. TERMS. Deposit £4/5/12 months of £1/38.
(b) The above Units with the B.S.R. MONARCH 3-SPEED CHANGER for ... £28/10/H.P. TERMS. Deposit £6/12/-. 12 months of £1/16/11.
EQUIPMENT OF THIS NATURE ENSURES "FIDELITY" REPRODUCTION ON BOTH
RADIO AND GRAM.

Home Constructors



YOU CAN ASSEMBLE FOR

H.P. Terms are shown below.

IT ONLY NEEDS CONNECTING

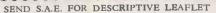
We are completely satisfied that this Tape Recorder, although supplied at a genuinely low price, provides absolute Fidelity Recordings and, in addition to being completely dependable, has a performance at least equal to recorders marketed at a far higher price. The actual assembly of the Tape Recorder is extremely simple and only involves a few connections. The Truvox Tape Deck and the Quality Amplifier are supplied tested and ready for use, and all that is required to complete the Recorder is to connect the two together (a connection chart is supplied for this purpose) and secure them by the screws provided into the Attache Case. The items illustrated and described below form the complete equipment.

WILL TAKE ALL STANDARD TAPES UP TO

• WILL PLAY THE NEW PRE-RECORDED TAPES

WILL PROVIDE 2 HOURS' PLAYING AT 3 in. or I hour at 7gin. per second

• INCORPORATES AN ELLIPTICAL P.M. SPEAKER 7×4in., with EXTENDED FREQUENCY RANGE





THE NEW TRUVOX MODEL TR7U TAPE DECK

THE NEW TRUVOX MODEL TR7U TAPE DECK. 3 Shaded-Pole motors. Drop-in Tape Loading. Push Button Control. Separate Push Button Brake. Fast forward and fast reverse. Silent drive eliminating Wow and Flutter. Half Track working and 2 speeds, 3fin. and 7fin. per sec. Positive Azimuth Adjustment. Overall size only 141 x 123in.



ACOS CRYSTAL MICROPHONE MODEL MIC.33.1

REEL OF NETIC RECORDING TAPE.

MODEL T.R.I./F. QUALITY AMPLIFIER

This amplifier has been expressly designed to meet the requirements of enthusiasts for fidelity reproduction, and in particular to CORRECTLY operate the above TRUVOX DECK. It is supplied complete with a matched Elliptical 3 ohm P.M. Speaker, it incorporates an efficient Tone Control arrangement and has a Magic Eye Level Indicator (Operative on Record). In addition it can be used as a general purpose Amplifier for high quality reproduction of gramophone records direct from a Gram Unit.



GUARANTEED FOR 12 MONTHS (B.V.A. VALVES 90 DAYS)

PRICE SUMMARY

WE WILL SUPPLY ALL FIVE UNITS LISTED ABOVE, i.e., THE COMPLETE BUT UNASSEMBLED RECORDER FOR £40/-/-. H.P. Terms: Deposit £10 and 12 monthly payments of £2/15/0 or in two parts as follows:-12 monthly DEPOSIT payments of PRICE

(a) TRUVOX MR. TR7U TAPE DECK MODEL TRIF AMPLIFIER WITH SPEAKER, 1,200ft. REEL OF TAPE £33 10 0

refunded if case is returned to us intact.

£8 10 0 See note below re packing charge

(b) ATTACHE CASE AS ILLUSTRATED ACOS CRYSTAL MICROPHONE 66 10 0 — —

NOTE: Please send 30/- to cover cost of packing, carriage and insurance. We will refund £1 if the packing case is returned to us intact.

EACH UNIT IS AVAILABLE SEPARATELY AS FOLLOWS:

CASH 12 monthly (a) TRUVOX Mk. TR7U TAPE DECK. £23 2 0 £5 17 0 £1 12 0 (b) AMPLIFIER MODEL TRIF WITH SPEAKER. £14 14 0 £4 16 6 18 4 (c) PORTABLE ATTACHE CASE £5 0 0 — — — (e) REEL OF TAPE 1,200ft. £1 15 0 — — (e) REEL OF TAPE 1,200ft. £1 15 0 — — Please Include £1 when ordering (a) or (c) for packing charge, this whole amount will be refunded if case is returned to us intact. PRICE DEPOSIT payments of

PORTABLE ATTACHE CASE

may be judged from the illustration opposite, is a neat, compact and attractively fin-ished case, being covered with maroon rexine and having an ivory coloured speaker escutcheon. It contains concealed pockets to accommodate the Microphone, Mains Lead and a spare 1,200ft. reel of tape

PRICE REDUCTION OF THE ASSEMBLED RECORDER

WE HAVE DISCONTINUED SUPPLY-ING WHOLESALERS. THE SAVING EFFECTED ENABLES US TO REDUCE THE PRICE AND OFFER IT COMPLETE AND READY 243'-/FOR IMMEDIATE USEFOR 24'-/-(Plus £1/10/- carriage and insurance. £1 is refunded when packing case is reinsurance. turned to us.)

H.P. terms. Deposit payments of £2/18/8. Deposit £11 and 12 monthly

PORTABLE MAKE YOUR OWN RECORD PLAYER!!!

WE OFFER THE PORTABLE CASE ILLUSTRATED

CONTAINING A 3-VALVE AMPLIFIER AND A P.M. SPEAKER FOR ONLY £7/17/6

(plus 10)- carriage and insurance—7/8 refunded on return of Packing Case). The Portable Case was originally used by Collaro Ltd. for their Microgram Record Player. It is robustly constructed and finished in good quality grey rexine. It will accommodate the COLLABO R.C.54
3-8FEED AUTOCHANGER and any make of 3-speed Single Record Player, all of which we can supply.
The Amplifier will operate with any high impedance Pick up and comprises a 3 VALVE A.C. MAINS design employing a 6K6 Output Valve for approx. 3 watts output. Tone and Volume Controls, combined with On-Off Switch are provided together with a coloured indicator.

indicato

Case and Amplifier are available separately.

(a) Portable case (plus 10/- carriage and insurance, 7/6 refunded).
(b) 3-VALVE AMPLIFIER, with P.M. SPEAKEE.
(Plus 3/- carriage and insurance.)



£3 7 6 £4 10 0



An "Alldry" Battery Portable of midget size, 6 lin. × 4 lin. × 3 lin. designed to cover medium wave-band 190-559 metres, with use of short trailer aerial.

suort vraiter aerial.
The simple design of this Receiver
is so arranged that either a 3-valve
set or a 2-valve (afterwards easily
converted to the 3-valve) can be
made.

made.
Consists of a T.R.F. circuit using a regenerative detector with H.F. stage and a high gain output pentode. Valve line up 1T4—1T4—DL94.
The 2-valve set can be completely built for \$24/3/8 (less case) and the 3-valve for \$25/3/- (less case). Each price includes valves, speaker and drilled chassis.

Send 2/- for the assembly instructions: they include simple and complete practical component layouts and diagrams.

STERN'S BARGAIN HAVE BOUGHT **FAMOUS** MODEL B3PP RADIOGRAM RADIO CHASSIS

A 6 VALVE 3 WAVEBAND SUPERHET with PUSH-PULL OUTPUT

Thousands of these successful and very popular Receiver Chassis have been sold for £15/15/- each.

WE CAN NOW OFFER THEM FOR £11/19/6

(plus 7/6 carriage and insurance), H.P. Terms. DEPOSIT £3 and 12 Monthly payments of 17/5.

GENERAL DETAILS

For use on A.C. Mains 100/110 Volts and 200/250 volts.

Employs the latest Valves 6BE6, 6BA6, 6A76, two 6BW6's in push-puil and 6X4 (or similar) Rectifier.

It has a Mains socket on the chassis for connection to Gram unit.

Incorporates extension speaker and Pickup sockets.

Overall size of Chassis is Ilin. × 7½in., 8½in. high.

Dial size 8½in. × 4½in.

(A Bronze coloured Dial Escutcheon is available for 4/6.

These Receivers Chassis have undoubtedly proved to be about the most popular and successful yet offered. They are designed to the most modern specification with great attention baving been given to the quality of reproduction which gives really excellent clarity of speech and music on both Radio and Gram.

THEY ARE THE IDEAL REPLACEMENT CHASSIS FOR THAT "OLD RADIOGRAM"

ALL CHASSIS ARE BRAND NEW and GUARANTEED FOR 12 MONTHS (B.V.A. VALVES 90 DAYS).

"PERSONAL SET" BATTERY ELIMINATOR

(2) Tone Control (3) Tone Control (4) A complete Kit of parts to build a Midge' and Radiol (5) and Radiol (5) and Radiol (6) and Radiol (7) approx. 69 volts at 10 mA and 11.4 volts at 250 mA. Size of assembled and guickly assembled and guickly assembled and guickly assembled and assembly instructions, 42/6.

In addition we can offer a similar COMPLETE KIT to provide approx. 90 volts at 10 mA, and 1.4 volts at 250 mA. Size of assembled unit 7in.×21in.×11in. Price 47/6.

A BULK PURCHASE ENABLES THIS SPECIAL PRICE REDUCTION OF THE FAMOUS

SHAFTESBURY PORTABLE AMPLIFIER



Suitable for home use and small Halls. Has matched inputs for both Record Players and Microphone. Also pro-vides for the "mixing" and "fading" of both Gram, and speech as request.



COMPRISING

A 4-Valve High Gain Amplifier for use on A.C. or D.C. mains. 200-230 volts with 5 watts output. Incorporating independent Volume Controls for Mike and Gram., either of which can be faded at will, a variable Tone Control and independent input sockets for Mike and Gram.

A Transverse Carbon microphone which obtains its polarizing current from the amplifier—no batteries are necessary.

A Sin. Goodmans P.M. Speaker with the "Ticonal" magnet for first-class reproduction.

(c) A

THE COMPLETE EQUIPMENT is all contained in the

PORTABLE CARRYING CASE £18'0

HIRE PURCHASE TERMS Having been reduced from £30/9/-. HIRE PURCHASE THE DEPOSIT £4/10/- and 12 monthly payments of £1/5/4 © in weight © Easy to CARRY © GENUINELY PORTABLE. illustrated leaflet containing free data is available on receipt of £ reduced from £30/9/-.

DUAL-CHANNEL PRE-AMPLIFIER and TONE CONTROL UNIT

Attractively finished in "Old Gold" and providing full control of BASS and TREBLE in conjunction with a main volume control.

It can be used with any amplifier and with any pickup, the range of frequency control provided by the unit affording ample compensation for all types of pick-ups and all natures of recordings, i.e., English, American and long-playing without recourse to pick-up correction. The extreme flexibility of the bass and treble control is such that the level of bass and treble can be act to suit any conditions irrespective of the volume output of the amplifier. Response characteristics are given in 12-watt amplifier advt. The unit means control is such that the level of bass and treble control is such that the level of bass and treble can be accommodated either on or away from the main amplifier, i.e. on the front panel of a cabinet or any other position. Frice including drilled chassis, valves (68N7 and 63D, 23/169. Complete assembly data are available separately for 1/s. Completely assembled and ready for use, 25/5/s.

THE IDEAL SET FOR USE IN CARAVANS, ETC. !!

A 5-VALVE 2-WAVEBAND SUPERHET RECEIVER OPERATED FROM A 6-VOLT BATTERY FOR ONLY \$6'17'6

6-VOLT BATTERY FOR ONLY 20' 11' 0

(plus 5). Carriage and Insurance).

These Receivers, which we have recently acquired by bulk purchase, are ex-British Ministry of Supply, and are new and unused. They are a two-wavehand Superhet with E.F. Stare, covering Short Wave 18 to 50 metres and Medium Wave 200 to 550 metres, fully calibrated on a clockface dial. A fill. loudspeaker is built in and the whole Chaesis is contained in a metal cabinet with lid and carrying handle which measures 12tin. x 74in. x 74in. w 74in. w



WILLIAMSON AMPLIFIERS BY GOODSELL

WILLIAMSON AMPLIFIERS BY GOODSELL
These Amplifiers hardly need enlarging upon, being sufficient to say that they have now become the accepted standard for quality reproduction by which all others are judged. Two Models are available:

MODEL G.W.12. These slightly lower H.T. voltage to produce 10-12 watts output but otherwise is built completely to specification.

Price \$27/10/= (Plus 7/6 Carriage H.P. Terms Deposit £6/17/6 and Insurance.)

THE MODEL P.F.A. TONE CONTROL UNIT
This Control Unit has established a reputation for its excellent quality of reproduction and ability to give adequate gain for any type of pick-up.

Price \$20/-/- (Plus 7/6 Carriage H.P. Terms. Deposit £5 and 12 months at £1/8/2.

SEND S.A.E. FOR ILLUSTRATED LEAFLETS

109 and 115 FLEET S' LONDON. E.C.4. Phone: CENTRAL 5812-3-4

SELENIUM RECTIFIERS

L.T. Types		H.T. Typa H.W.	
2/6 v. 1 a.h.w	1/9	120 v. 40 mA	3/11
6/12 v. 3 a.h.v	v. 2/9	250 v. 50 mA	5/9
		250 v. 80 mA	7/9
F.W. Bridge Ty	/pas	250 v. 150 mA.	9/9
6/12 v. 1 a	5/9	RM4 250 v. 250 m A	11/3
6/12 v, 2 a	8/9	300 v. 275 mA.	12/11

CO-AXIAL CABLE. 75 ohms lin., 7d. yard. Twin screened feeder, 10s. yard.

SILVER MIGA CONDENSERS. 5, 10, 15, 20, 25, 30, 35, 50, 100, 120, 150, 180, 200, 230, 300, 330, 400, 470, 500, 1,000 ptd. (.001µF), .002 mtd. (2,000 ptd.), All at 31. eaca, 3/3 dozen one type.

DIAL BULBS, M.E.S., 8 v. 0.15 a., 6/9 doz.; 6.5 v. 0.3 a., 6/9 doz.; 4 v. 0.3 a., 6/- doz.

ELECTROLYTICS (current production). NOT ex Govt.

Tubular Types		Can Types	
\$\textit{Suff 450 v} \\ \text{S mfd. 500 v} \\ \text{16\text{\$\mu F\$ 350 v} \\ \text{16\text{\$\mu F\$ 350 v} \\ \text{16\text{\$\mu F\$ 350 v} \\ \text{16\text{\$\mu F\$ 500 v} \\ \text{32\text{ mfd. 500 v} \\ \text{25\text{\$\mu F\$ 500 v} \\ \text{25\text{\$\mu F\$ 500 v} \\ \text{50\text{\$\mu F\$ 50 v} \\ \text{50\text{\$\mu F\$ 50 v} \\ \text{100\text{ mfd. 25 v} \\ \text{100\text{ mfd. 25 v} \\ \text{can Types} \\ \text{s mfd. 450 v} \\ \text{16\text{ infd. 500 v} \\	1/9 2/3 2/3 3/9 3/9 3/9 5/9 4/11 1/3 1/3 2/3 1/9 2/3	$\begin{array}{c} 16 \text{ mfd. } 350 \text{ v} \\ 16 \mu \text{F } 450 \text{ v} \\ 24 \mu \text{F } 350 \text{ v} \\ 32 \mu \text{F } 350 \text{ v} \\ 32 \text{ mfd. } 450 \text{ v} \\ 64 \text{ mfd. } 450 \text{ v} \\ 100 \text{ mfd. } 450 \text{ v} \\ 8-8 \mu \text{F } 450 \text{ v} \\ 8-8 \mu \text{f. } 450 \text{ v} \\ 8-16 \mu \text{F } 450 \text{ v} \\ 16-16 \mu \text{F } 450 \text{ v} \\ 16-32 \mu \text{F } 350 \text{ v.} \\ 32-32 \mu \text{F } 350 \text{ v.} \\ 32-32 \mu \text{F } 450 \text{ v.} \\ \end{array}$	1/11 2/9 2/11 2/11 4/9 4/9 3/6 4/9 2/11 4/11 4/9 4/9 5/11
8 mfd. 450 v	2/3 3/9	32-32μF 350 v.	4/9

VOLUME CONTROLS with long spindles, all values, less switch, 2/9; with S.P. switch, 3/9.

WIRE WOUND POTS: 20 ohms, 500 ohms, 5K, 20K, 100K (medium length spindles), 2/9. 220 ohms, 2K, 10K, 20K, Preset type, 1/9 each.

VIBRATORS. Wearite 12 v. 4 pin. Non synchronous, 6/9. Oak 2 v. 7 pin, synchronous 7/9.

			5/9
		PAPER CONDEN	
		6-6 mfd. 450 v.	
		8 mfd. 500 v	
4 mfd. 1,000 v.	4/3	8-8 mfd. 500 v.	5/11
4 mfd. 1,500 v.	4/9	15 mfd. 500 v	7/9
4 mfd. 400 v. plus	2 mfd.	250 v	1/11
	. 1	DEDOG 1 1 1I	

EX. GOV. UNITS, type RF26 in original sealed cartons 39/6. Transmitter Receivers type TR9D complete with all valves 45/-, carr. 6/6.

M.E. SPEAKERS. All 2-3 ohms, 8in. R.A. field' 600 ohms, 11/3. 10in. R.A. field, 1,500 ohms, 23/9-10in. R.A. field, 1,000 ohms, 23/9.

MANUFACTURERS SURPLUS
TRANSFORMERS
TRANSFORMERS Fully shrouded upright. Primary 200-230-250 v. Sec. 425-6-425 v. 150 m.a. 6.3 v. 3 a. 5 v. 3 a. 37/2.

GOODMANS 31in. P.M. SPEAKER (ex. equip.), with battery pentode trans., 12/9.

HEAVY DUTY BATTERY CHARGER

For normal 200/250 v. A.C. mains input. To charge 12 v. battery. Variable charge rate of up to 10 amps. Fitted Meter and Fuses. Guaranteed 12 months. Carr. 10/-. £6/19/6.

OIL FILLED BLOCK CONDENSERS
Bryce 11-7 mfd. 500 v. New unused Govt. surplus, only 5/9 each.

H.T. ELIMINATOR AND TRICKLE CHARGER KIT with louvred crackle finished case. Mains input 200-250 v. Output 120 v. 40 mA., and 2 v. a. Price with circuit, 29/6. Or in working order, 37/6.

TRANSFORMERS R.S.C.

IMPREGNATED

FULLY GUARANTEED, I	NTERL	EAVED AND IMPREGNAT	ED
MAINS TRANSFORMERS		FILAMENT TRANSFORMERS	
Primaries 200-230-250 v. 50 c/s.		Primaries 200-250 v. 50 c/s.	
FULLY SHROUDED UPRIGHT MOU		6.3 v. 1.5 a 5/9 0-4-6.3 v. 2 a	7/9
250-0-250 v. 60 mA. 6.3 v. 2 a., 5 v. 2		6.3 v. 3 a 8/11 0-2-4-5-6.3 v. 4 a.	
Midget type, 2½-3-3in	17/6	0.3 V. 6 a	17/6
350-0-350 v. 70 mA., 6.3 v. 2 a., 5 v. 2 a.		12 7, 5 4, 01 23 7.	
250-0-250 v. 100 mA., 6.3 v4 v4 a., c		6.3 v. 2 a 7/6 1.5 a	17/6
0-4-5 v. 3 a. 250-0-250 v. 100 m \(\), 6.3 v. 4 a., 5 v. 3 a	26/9	CHARGER TRANSFORMERS	
250-0-250 V. 100 m 1., 0.3 V. 4 a., 5 V. 5 a.	23/9	All with 200-230-250 v. 50 c/s. Primaries: 0-9	-15 v.
250-0-250 v. 100 mA., 6.3 v. 6 a., 5 v. 3 for R1355 conversion		11 a., 11/9; 0.9.15 v. 3 a., 0.3.5.9-17 v. 4 a.,	18/9.
300-0-300 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a		0-9-15 v. 5 a., 19/9; 0-9-15 v. 6 a., 23/9.	
300-0-300 v. 100 mA., 6.3 v. 4 a., 5 v. 5 a		ELIMINATOR TRANSFORMERS	
		Primaries 200-250 v. 50 c/s. 120 v. 40 mA.	7/11
0-4-5 v. 3 a. 350-0-350 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a	23/9	130 v. 50 mA., 6.3 v. 3 a	14/9
350-0-350 v. 100 mA., 6.3 v4 v., 4 a., c	.t	120 v. 40 m.A., 5-0-5 v. 1 a.	14/9
350-0-350 v. 100 mA., 6.3 v4 v., 4 a., c 0-4-5 v. 3 a.	26/9	90 v. 15 mA., 6-0-6 v., 250 mA	9/11
350-0-350 v. 150 mA., 6.3 v. 4 a., 5 v. 3 a	33/9	OUTPUT TRANSFORMERS	
350-0-350 v. 150 mA., 6.3 v. 2 a., 6.3 v. 2	a.,	Midget Battery Pentode 66:1 for 3S4, etc	3/6
5 v. 3 a. 425-0-425 v. 200 mA., 6.3 v. 4 a., c.t., 6.3	33/9	Small Pentode, 5,000Ω to 3Ω	3/9
425-0-425 v. 200 mA., 6.3 v. 4 a., c.t., 6.3	v.	Standard Pentode, 5,000Ω to 3Ω	4/9
4 a., c.t., 5 v. 3 a., suitable Williams Amplifier, etc.	son	Standard Pentode, 8,000 to 3\Omega	4/9
Ampliner, etc.	49/9	Battery Pentode, 10,000 ohms to 3 ohms	4/9
450-0-450 v. 250 mA., 6.3 v. 6 a., 6.3 v. 6 5 v. 3 a.	a., 69/6	Multi-ratio 40 mA. 30:1, 45:1, 60:1, 90:1.	
TOP SHROUDED DROP THROUGH	TVPF	Class B Push-Pull	5/6
250-0-250 v. 70 mA., 6.3 v. 2.5 a		Push-Pull 8 Watts 6V6 to 3 ohms	8/9
260-0-260 v. 70 mA., 6.3 v. 2 a., 5 v. 2 a.		Push-Pull 10-12 Watts 6V6 to 3Ω to 15Ω ,	
350-0-350 v. 80 mA., 6.3 v. 2 a., 5 v. 2 a		sectionally wound	16/9
250-0-250 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a	22/9	Push-Pull 10-12 Watts to match 6V6 to	4010
300-0-300 v. 100 mA., 6.3 v4 v. 4 a., c	.t.,	3-5-8 or 15Ω Push-Pull 15-18 Watt, sectionally wound,	16/9
U-4-5 v. 3 a	23/9	6L6, KT66, etc., to 3 or 15 ohms	21/9
350-0-350 v. 100 mA., 6.3 v. 4 a., c.t., 5	V	Push-Pull 20 Watt high-quality sectionally	21/3
3 a	22/8	wound, 6L6, KT66, etc., to 3 or 15Ω	47/9
350-0-350 v. 100 mA., 6.3 v4 v. 4 a., c	3.L.,		,.
0-4-5 v. 3 a	23/9	SMOOTHING CHOKES 250 mA.,3 H., 100 ohms	44/0
350-0-350 v. 150 mA., 6.3 v. 2 a., 6.3 v. 2 5. v. 3 a.	29/11		11/9
350-0-350 v. 150 mA., 6.3 v. 4 a., 5 v. 3 a	29/9	100 mA., 10 H., 150 ohms potted	9/9
E.H.T. TRANSFORMERS, 2,500 v. 5 m	A.,	100 mA., 10 H., 200 ohms	8/9
2-0-2 v. 1.1 a., 2-0-2 v. 1.1 a., for VCR		80 mA., 10 H., 350 ohms	5/6
VCR517		60 mA., 10 H., 400 ohms	4/11
	,		

THE SKY FOUR T.R.F. RECEIVER





A design of a 3-valve 200-250 v. A.C. A design of a 3-valve 200-250 v. A.C. Mains receiver with selenium rectifier. For inclusion in either of cabinets illustrated above. It employs valves 6K7, SP61, 6F66, and is specially designed for simplicity in wiring. Sensitivity and quality is well up to standard. Point-to-point wiring is well up to standard. Point-to-point wiring diagrams, instructions and parts list, 2/3. This receiver can be built for a maximum of 24/19/6 including cabinet. Available in brown or cream bakelite, or veneered walnut.

P.M. SPEAKERS. All 2-3 ohms. 61 in. Plessey with 5,000 ohm output transformer, 16/11. 8in. Goodmans, 19/9. 10in. R.A., 26/9. 10in. Rola with Trans., 29/6.

R.S.C. BATTERY CHARGER KITS. For mains input 200-250 v. 50 c/s. To charge 6 v. accumu-



lator at 2 amps, 25/9.
To charge 6 v. or 12 v.
battery at 2 a., 31/6.
To charge 6 v. or 12 v.
battery at 4 a., 49/9.
ABOVE KITS CONSIST

OF GREEN CRACKLE
LOUVRED STEEL
CASE, MAINS TRANSFORMER, FULL WAVE METAL RECTIFIER,
FUSES, FUSE-HOLDERS AND CIRCUIT. FUSES, FUSE-HOLDERS AND CIRCUIT.
Any type assembled and tested for 6/9 extra.

BATTERY CHARGER

R.S.C. 6 v. or 12 v. B For normal A.C. mains input 200-230-250 v., 50 c/s. Selector panel for 6 v. or 12 v. charging. Variable charge rate of up to 4 AMPS. Fused, and with ammeter. Well ventilated metal case with attractive crackle finish. Guaranteed for 12 months, 69/6. Carr. 2/6.



EX GOVT. MAINS TRANSFORMERS

-	All 230 v. 50 c/s, input.	
ora 1	8.8 v. 4 a.	9/9
	48 v, 1 a	9/9
	300-0-300 v, 80 mA, 5 v, 3 a,	8/11
	278-0-278 v. 100 mA	8/9
-	Carriage on following types 5/-	extra
	0-11-22 v. 30 a	72/6
No.	16-18-20 v. 35 a	79/6
	7.7 v. C.T. 7 amps., 4 times	25/9
	460 v. 200 mA., 6.3 v. 5 a	27/9
300-0-3	300 v. 150 mA., 610-0-610 v. 150 mA.,	
1,200	0 v. 250 mA	29/6
	C.T. 150 mA, 4 v. 5 a., 6.3 v. 6 a.,	
	7. 0-6 a., 4 v. 6 a., 4 v. 6 a., 4 v. 3 a.,	
	3 a., 5 v. 2 a	22/9
325-0-3	325 v. 150 mA., 6.3 v. 4-6 a., 5 v. 2-3 a.	29/9

EX GOVT. AUTO TRANSFORMERS

	15-10-5-0-195-215-235 v. 500 watts	27/9
	Double wound 10-0-200-240 v. to 10-0-275-	
	295-315 v. 1,000 watts	69/6
ı,	Double wound 0-110-240 v. to 0-130-140-	
	150-160-170 v. 1,500 watts	69/6
	Carriage on any of above 5/- extra.	

EX GOVT. SMOOTHING CHOKES

= 11 00011011100	
250 mA., 10 H., 50 ohms	14/9
250 mA., 10 H. 100 ohms	14/9
	8/9
150 mA., 10 H., 50 ohms	
100 mA., 10 H., 100 ohms, Tropicalised	6/9
100 mA., 5 H., 100 ohms, Tropicalised	3/1
50 mA., 50 H., 1,000 ohms, Potted	8/1
90/100 mA., 10 H., 100 ohms, Potted	8/9
50 mA., 5-10 H.	2/9
L.T. type 1 amp.	2/9

CHASSIS

18 s.w.g. undrilled aluminium amplifier type (4-sided).

14in.×10in.×3in. 7/11 16in.×10in.×3in. 8/3 18 s.w.g. aluminium receiver type.

6in.×3§in.×1§in. 1/11 7§in.×4§in.×2in. 2/9 10in.×5§in.×2in. 3/3 11in.×6in.×2§in. 3/11

16 s.w.g. al receiver type. aluminium

12in. × 8in. × 21in. 5/3 16in. × 8in. × 2½in. 7/6 20in. × 8in. × 2½in. 8/11 16 s.w.g. aluminium amplifier type, 4-sided.

12in. × 8in. × 21in. 7/11 16in. × 8in. × 21 in.10/11 20in. × 8in. × 21 in.13/6 14in. × 10in. × 3in.13/6

R.S.C. HIGH FIDELITY 2.5 watt AMPLIFIER

A NEW DESIGN FOR 1955 HIGH GAIN "PUSH PULL OUT-PUT". BUILT-IN PRE-AMP. TONE CONTROL STAGES. INCLUDES 7 valves, sectionally wound output transformer, block paper reservoir condenser, and reliable small components. AN INPUT OF ONLY 20 millivolts IS REQUIRED FOR FULL OUTPUT. THIS MEANS THAT ANY TYPE OF MICRO-PHONE OF DIEGOTOR HOLES WITH A PROPERTY OF THE PROP PHONE OR PICK-UP IS SUITABLE Two separate inputs controlled by separate volume controls allow simultaneous use of "Mike" and Gram., or Tape and Radio, etc., etc. Individual controls for Bass and Treble "lift" and "cut".

Six negative feedback loops giving total of 24 D.B. Frequency response ± 3 D.B. 30-20,000 c/s.

Hum level 66 D.B. down. Certified total harmonic distortion of only 0.35% measured at 10 watts. Comparable with the very best designs. SUITABLE FOR SMALL HOMES OR LARGE HALLS, CLUBS, GARDEN PARTIES, DA DANCE HALLS, etc., etc. For ELECTRONIC ORGAN OR GUITAR. For STANDARD OR LONG PLAYING RECORDS. Size 12 × 10 × 9in. For mains A.C. 200-250 v. 50 c/s. Power consumption 175 watts. Outputs for The kit is 3 and 15 ohm speakers. complete in every detail. Chassis is

fully punched. Easy to follow point-to-point wiring diagrams are supplied. EXTRA HIGH SENSITIVITY, HIGHEST QUALITY for Or assembled ready for use 50/- extra. 9 GNS.

H.P. Terms on assembled units. Deposit 26/- and 12 monthly payments of £1. Plus carr. 10/-. Terms to include cover, mike, speakers, etc., on request. Cover as illustrated if required, price 17/6 extra.

A PUSH PULL 3-4 WATT HIGH GAIN ASSEMBLED AMPLIFIER FOR £3/19/6. For mains input 200-250 v. 50 c/s. Complete kit of parts including point-to-point wiring disgrams and instructions. Amplifier can be used with any type of feeder unit or pick-up. This is not A.C./D.C. with "live" chassis but A.C. only with 400-0-400 v. Trans. Output is for 2-3 ohm speaker. Supplied ready for use, £3/19/6. Full descriptive

H.M.V. LONG PLAYING RECORD TURNTABLE COMPLETE WITH CRYSTAL FICK-UP (SAP-PHIRE STYLUS, Speed 33 r.p.m. BRAND NEW, CARTONED. Only £3/19/6 (approx. half price). Carr. 5/- (for 200-250 v. A.C. Mains).

R.S.C. A7 3-4 WATT QUALITY AMPLIFIER R.S.C. A7 3-4 WATT QUALITY AMPLIFIER A highly sensitive 4-vaive amplifier using negative feedback and having an excellent frequency response. Fre-amplifier and Tone Control stages are incorporated with separate Bass and Treble controls giving full tone compensation for Long Phyling records. Suitable for any kind of pick-up including latest high fidelity types. H.T. of 250 v. 20 m.a. and L.T. 6.3 v. 1.a. available for supply of Radio Feeder Unit, etc. ONLY 40 millivolts input required for full output. Fully isolated chassls with usacplate. For A.C. mains 200-250 v. 50 cycles. Complete kit of parts with point-to-point wiring diagrams and instructions. Only £3/15/-..

R.S.C. 4-5 WATT HIGH GAIN AMPLIFIER TYPE A5



A highly sensitive 4-valve quality amplifier for the home, small club, etc. Only 30 millivolts input is required for full output so that it has been also also the with the latest high control of the with the latest high relative to the with the latest high relative highly and practically all mikes. Separate Bass and Treble controls are provided. These give full long playing record equalisation. Hum level is negligible being 71 D.B. down. 15 D.B. of negative feedback is used. H.T. of 30 v. 25 mA. and L.T. of 6.3 v. 1.5 a. is available for the supply of a Radio Feeder Unit, or Tape Deck pre-amplifier. For A.C. mains input of 200-230-250 v. 50 of s. Chassis is not alive. Kit is complete in every detail and includes fully punched chassis (with baseplate, with green brackle finish, and point-to-point wiring diagrams and instructions. Exceptional value at only \$4(15)-r, or assembled ready for use 25/- extra, plus 3/6 carr. Output for 3-ohm speaker.

BRAND

With crystal pick-up and separate sapphire point stylit for standard or long playing records. Plays ten 7in., 10in. or 12in. intermixed. Supplied in sealed cartons with or 12in. intermixed. Supplied in scaled cartons template and operating instructions. Only £9/19/6, 5/- carr., or 2 gns. deposit and 11 monthly payments 1 COLLARO HIGH FIDELITY MAGNETIC PICK-UPS High impedance type. Limited number, brand new, boxed and perfect at fraction of normal price. Only 35/-.

DEFIANT RECORD PLAYING TURNTABLE COMPLETE WITH MAGNETIC PICK-UP. Pick-up is high impedance type. Unit is housed in a beautiful walnut veneered cabinet of attractive design. For all standard records (78 r.p.m.). Limited number. Brand new, cartoned, £5,19 8. Carr. 7/6.

ACOS HIGH FIDELITY CRYSTAL MICRO-PHONES. Type 22-2. Complete with table stand. Normal price 4 gns. Limited stocks, brand new, boxed, £2:19:6.





R.S.C. MASTER INTERCOMM. UNIT, with provision R.S.C. MASTER INTERCOMM. UNIT, with provision for up to 4 "Listen-Taik Back Units" individually switched. A high gain amplifier enables speech and other sounds emanating from the rooms containing remote control units to be heard at the master control. Supplied with walnut veneered wood or brown bakelite cabinet. Mains input is 200-250 v. 50 v.s. H.T. line 300 v. CRASSIS 18 NOT "ALIVE." Ideal for use as "Baby Alarm." Sound amplification of wasts. Frice only 27/15f-. "Listen-Taik Back Unit" in bakelite or walnut veneered cabinet. San be sumited at 35% cach. n be supplied at 35/- each

All parts for an "All Dry" Battery Ellminator, Complete with case. Completely replaces 1.4 v. and 90 v. batteries where normal mains supply of 200-220 v. 50 c/s. is available. Price with circuit, 38/9. Or ready for use, 45/8. Size of unit 5\frac{1}{2} \times 4\frac{1}{2} \times 2\frac{1}{2} \times 1.0 \text{ and } \times 2.0 \text{ and } \

ALL DRY RECEIVER BATTERY SUPER-SEDER KIT

thereby covering latest low consumption types.

BATTERY SET CONVERTER KIT. All parts for converting any type of battery receiver to all mains. A.O. 200-250 v. 50 9.06 %. Kit will supply fully smoothed H.T. of 120 v. 90 v. or 60 v. at up to 40 mA., and fully smoothed L.T. of v. at 0.4 v. at 0.4 x. to 1 a. Frice complete with circuit and instructions only 48/9. Supplied ready for use for 8/9 extra.

LEEDS, 2. THE 32

Terms C.W.O. or C.O.D. No C.O.D. under £1. Postage 1/- extra under 10/-, 1/- extra under £2, 2/6 extra under £3, Full Price List 6d. Trade List 5d.

Open to Callers: 9 a.m. to 5.30 p.m. Saturday until 1 p.m.

NEW B.S.R. MONARCH 3-SPEED R.S.C. A3 10 WATT "PUSH-PULL" HIGH MIXER AUTO-CHANGERS. FIDELITY AMPLIFER

With Self-Contained Pre-amplifier and Tone Control



Large safety factors in every component A.C. and H.T. fuses, punched chassis with baseplate, screened Input plugs, 6 valves, and with easy-to-folic w point-to-point wiring diagrams. Everything supplied to last nut. Two independent inputs are provided with two associated independent volume controls so that programmes can be mixed together if desired, such as microphone announcements superimposed on a musical programme, or two independently controlled microphones, or even just gramo-phone/reddie, fadding over from one to the other. Varirble base lift and out with variable treble lift and cut too controls are fitted, giving full long playing record equalisation for uncorrected pick-ups. They are also provided so that the user can alter the tonal value to suit his personal taste and surroundings. Torninals are provided for 3 olum and 15 olum loudelpeakers.

nit.

Six Negative Feedback Loops.

130 millivolts input only required for full output.

Frequency response 50-20,000 cycles.

Negligible hum and distortion.

For A.C. mains input 200/230/250 v. 50 c/s.

COMPLETE Kit of Parts 7 GNS. (carriage 7/6). Supplied, assembled and tested for 45 - extra. Cover as for A4 amplifier 17/6 extra if required. HP. TERMS on assambled units. Deposit 23/8 and 9 monthly payments 21/-.

FOUR-STAGE RADIO FEEDER UNIT.
Design of a HIGH FIDELITY L. and M. wave T.R.F.
Unit with self-contained heater supply and thorough H.T.
decoupling. Only 250-400 v. 18-20 mA. H.T. required
from main supplifier. Three valves and Low Distortion.
Germanium Blods Detector. Flat topped response characteristic. Louised H.F. coils. Two variable Mu controlled
H.F. stages, 3 gang condenser tuning. Cathode follower
output stage. Switch position for Gram. and Gram. input
and output societs. Performance comparable with the,
best in Feeder Units. For A.O. mains 200-230-250 v.
operation. Biza 11-6-7-jin. Histartaion, full set of easy-tofollow wiring diagrams and instructions and individually
priced parts list 2/3f. This unit can be built for only 28/15/including Dial and Drive Knobs and every tem required.

W.B. "STENTORIAN" HIGH FIDELITY P.M-SPEAKERS. HF1012, 10 watts, 15 ohm (or 3 ohm) speech coll. Where a really good quality speaker at a low price is required we highly recommend this unit with an amazing performance, £3/17/6.

MICROPHONES, Crystal, hand or Desk fidelity Acos, 50/-. Stand type with base and adjustable stem, £6/19/6. Both suitable for use with our amplifiers.

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Built into a strong wooden cablact 15in. x
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If required we shall be pleased to align this unit for constructors not possessing the necessary equipment for a charge of 7/6. N.E.—Valve line-up is 6AM6, 12AH8, 2-6BA6 and 6ALS. Chassis measures only 6/8 × 5/8 × 1/1m.

Demonstrations at 18, Tottenham Court Roadil

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This kit has been based on the booklet by Data Publications, price 2/- post free. With each booklet is enclosed our individually priced parts list. The construction and parts list. The construction and alignment of this tuner are no more difficult than a normal medium wave tuner. It is highly sensitive and free from drift. Incorporates 4 valves type 6AM6 and 2 specially graded G.E.C. Crystals. The kit worklood included will be charged with

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A three-valve plus metal rectifier receiver. A.C.
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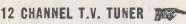


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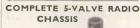
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LOUDSPEAKERS	211

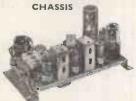
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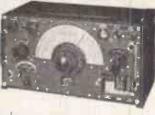
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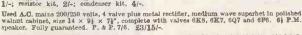
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Standard adaptable type, 230 v.
450 w., 1/8 ea. Morphy Richards, replacement type, 3/9 ea. H.M.V.

- manage		onsector Ald	1 Franti
		CATIONS	
Midget	Radio	Construct	lon
Manual			3/6
Radlofolo	der "F"	An Inexpens	ive
tane n	ecorder .		2/8
Radlofold	ler "B"	A practi	cal
Oscillo	scope		2/6
Radlo Ca	siculations	Manual .	3/6
		Manual	
		on Laborate	
Manual			2/6
	4		

	Eas
L.E.M. Silver Mica 1,000pF 109	% 31
L.E.N. Silver Mica 100 pf. 5%	
T.C.C. Silver Mica 50 pF. 10%	
L.E.M. Sliver Mica 300 pF. 10	
Bunts Silver Mica 374 pF. 1%	
L.E.M. Sliver Mica 350 pF. 10%	
L.E.M. Silver Mica 25 pF. 20%	
T.C.C. Silver Mica 1,000 pF. 10	% 31

Tubular Condenser .5 mfd. 500 v. 6d. Yaviey Switch 1 pole 8 way 11in.

Eric Mains Dropper 725 ohms Eric Mains Dropper 725 ohms
71% tapped 1 1/6
Dubliler Moulded Mica AP9 .005 414.
HF Pile Wound Chokes 1/7
Vibrator Clips 5d.
Dubliler Nitrogol B45 12 mfd.
350 v. 5/6
International Octal Valve Holders
Paxolin Vitreous Enamelled 20 w. 5,000 ohms 1/7

20 w. 5,000 ohms ... 1/
B.S.R. MONARCH AUTOMATIO
RECORD CHANGER
These units will autochange on all three speeds, 7in., 10in. and 12in.
They play MIXED 7in., 10in. and 12in.
They play MIXED 7in., 10in. and 12in. records.
They have separate sapphire for L.P. and 70 r.p.n. which are moved into position by a simple switch.
Minimum baseboard size required 14in.x12\$in., with height above 54m. and height below baseboard 2\$in. A bulk purchase enables us to offer these BRAND NEW UNITS at this exceptional price. These units are beautifully finished in cream enumel with cream bakelite arm.
COMPLETE WITH FULL INSTRUCTIONS. 29/19/6.

AMERICAN INDICATOR UNIT TYPE BC929A
Brased new incorporating 31n. tabe 3BPL new incorporating 31n. tabe 3BPL new incorporating 42n. August 1987 (1998) 1998 (1998)

Standard 11in, Brown Knobs, per

8tandard 1 in. Brown Knobs, per dozen 5/6
Zenith Dropper 910 ohms, each 2/6
Rakelite case. Double coll Buzzers, each. 2/6
Erie Dropper, 1,340 ohms, 150 ohms, each 2/6
Box. 4 BA Nuts and Bolts, each 1/2
Herralve Transformers Ex. Equip, each 2/2
Hand Microphone Bakelite switch in handle, each 7/6
High to low Impedance headphone units (insert in lead) each 3/6
Rubber grommets, assorted, per dozen 64
Bakelite Needle cups, each 1/6
Bakelite Needle cups, each 1/6
Bakelite Needle cups, each 1/6
Mains Switch, 2 P., 3 B., 3 W., each 1/6
Yaxieg Switch, 3 P., 3 B., 3 W., each 1/6
Yaxieg Switch, 1 Pole, 9 Way, each 1/9
Waxed Carton, 8 MFD., 450 v., each 1/9
Waxed Carton, 8 MFD., 450 v

TERMS: Cash with order or C.O.D. Postage and Packing charges extra, as follows: Orders value 10/- add 9d.; 20/- add 1/-; 40/- add 1/6; £5 add 2/- unless otherwise stated. Minimum C.O.D. fee and postage 2/3.

MAIL ORDER ONLY

MAINS TRANSFORMERS 3-WAY MOUNTING TYPE

MTI Primary: 200-220-240 v.
Secondaries: 275-0-275 v. 80 mA, 0-6.3 v. 4 amp.
0-5 v. 2 amp. Both tapped at 4 v. 17/6 ca MT2 17/6 ea.

PENCIL RECTIFIERS K3/25 5/8: K3/40, 7/6: K3/45, 8/2; K3/50, 8/8; K3/60, 9/8-K3/100, 14/8.

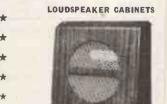
VOI DELI	
- {Each	Each
8 x 8 mfd. 450 v 4/-	32 x 32 mfd, 450 v 6/11
8 x 16 mfd. 450 v 4/-	32 x 32 x 8 mfd. 350 v. 5/6
8 x 24 mfd. 350 v 3/-	32 x 32 mfd. 350 v.
8 x 32 mfd. 475 v 3/9	25 m/d. 25 v 5/
12 x 4 mfd. 450 v 2/-	60 mfd. 450 v 2/9
16 m/d. 450 v 3/-	64 mfd. 350 v 2/-
16 × 8 mfd. 350 ▼ 4/-	Dubilier (B.R. Range):
16 × 16 mfd. 350 v 3/3	BR 850. 8 mfd. 500 v. 2/9
16 × 16 × 8 mfd. 350 v. 3/6	BR. 1650. 16 mfd. 500 v. 3/3
20 x 20 mfd. 500 v 4/9	BR.2050, 20 mfd. 500 v. 3/6
24 mfd. 450 v 2/9	8×8 mfd. 500 v 4/-
24 × 16 m(d. 350 v 3/6	BR.501, 50 mfd, 12 v 1/9
32 mfd. 450 v 3/-	
32 x 8 mfd. 350 v 3/6	16 x 16 m/d. 500 v 5/-
32 x 16 mfd. 350 v 4/6	16×8 mfd. 500 v 4/9

OSMOR COIL PACKS

Type H.O., 48/- each. Type L.M., 40/- each. Type T.B., 50/each. Type TRF, 40/- each.

SPRAGUE CONDENSERS

.05 mfd., 500 v.; 0.1 mfd., 1,000 v.; .1 mfd., 350 v.; .02 mfd., 750 v. All 9/- doz.



* * * * * * *

This attractive wainut finished cabinet is available for 6 in. or 8 in. speaker units. Metal speaker fret, complete with back and rubber feet.

6jin. type: Measures 8jin. × 8jin. × 4jin. at base-Price 16/6 each.

Sin. type: Measures 10in.×10in.×5in. at base. Price 20/6 each.

* * * * * LOUDSPEAKER UNITS

| Rola 51n. Speaker | 17/6 ea | 17/6 ea | 17/6 ea | 17/6 ea | 18/6 ea | 18/6

CHASIS

Aluminium Undrilled with Reinforced
Cornera. Available in the following
sizes.

6in. x 4in. x 24in. 4/6 ca.
8in. x 6in. x 24in. 7/13 ca.
10in. x 7in. x 24in. 7/13 ca.
10in. x 7in. x 24in. 9/6 ca.
14in. x 8in. x 24in. 9/6 ca.
14in. x 8in. x 24in. 1/2/c ca.
14in. x 8in. x 24in. 1/2/c ca.
All are four sided—ideal for radio receivers — amplifiers — powerpacks, etc.

PERSPEX IMPLOSION GUARDS Incorporates escutcheon and filter, 12in. type, 11/6 ea.; 16in. type, 14/6 ea.

EX GOVERNMENT GOVERNMENT AND SURPLUS CONTROLS

SURPLUS CONTROLS
This popular range is suitable for all Television constructors, etc. Keep process down when building the "Argus" or "Simplex" receivers. Available 500 R. 000 G. 1,500 G double type, 2 KG, 5 KG, 10 KG, 20 KG, 100 KG; ½ meg. G, ½ meg. G, ½ meg. G, ½ meg. G, 2 meg. G, 2 cach.



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4½ v. Heavy Duty Bell Battery. Size 6½ x 4½ x 2½in. 4/6 72 v. H.T. 1.5 v. L.T. Size 6 x 5 x 1½in. 5/6
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All batteries sealed and unused. All plus 1/6 post and pkg. Special
reduction for quantities.

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		doz.
	16 mfd. 375 v., 2/- each	
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	Condenser Clips for above	
	BIAS CONDENSERS	10//
	50 mfd, 12 v. Single Hole Fixing I/- each 100 mfd, 6 v. Tag End 10d. each	9/-
	100 mfd, 25 v, Tag End 1/3 each	
	BLOCK PAPER CONDENSERS 4 mfd, 400 v. D.C. 3/6 each. Many other types in stock. Your engi	inian
	Tillia, 400 V. D.C. 3/6 each, I fally belief types in stock. Tour end	411 IC3

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4 mfd. 400 v. D.C. 3/6 each. Many other types in stock. You	r enq	uiri	es
invited.			
LARGE ASSORTMENT OF TUBULAR CONDENSERS		6	1-
LARGE ASSORTMENT OF TUBULAR CONDENSERS MIDGET MICA CONDENSERS0001, .0002, .0003, .0004,	.0005	- 5	/-
200 Assorted Moulded Micas. Popular Values	£2	10 -	0
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200 Assorted Carbon Resistors: 1, 1 and I watt. Good selection	£I	10	0
SPECIAL OFFER		_	_

0.1 mfd. 12,000 volts test Mansbridge Condensers. Height 6½in. Width 3½in. Depth 2½in. Fixlng Centres 4in. Plus 1/- post ... 5/6 PAXOLIN SHEET

 $18 \times 4 \times 1/16$ in., 1/- each; $10 \times 10 \times 1/16$ in., 1/6 each; $20 \times 10 \times 1/32$ in., 1/6 each; $20 \times 10 \times 1/16$ in., 3/- each. RESISTORS

Carbon 1 watt 2/6; 1 watt 3/-; 1 watt 4/-; 2 watt 6/- per doz.
WIRE WOUND AND VITREOUS. 5 watt 1/6; 10 watt 2/6; 15 watt

WIRE WOUND AND VITREOUS. 5 watt 1/6; 10 watt 2/6; 15 watt 3/-; 20 watt 3/6 each.
HIGH STABILITY. ½ watt 5% 6d.; ½ watt 5% 9d.; 1 watt 5% 1/3 each.
A few values in 19% and 2% still available.
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Incorporating high "Q" technique using the New Ferrite rod. Made possible by simple conversion of an ex-Govt. Hearing Aid.

Technical Details. A Germanium Diode Detector circuit followed by the existing 3 valve Amplifier, giving adequate amplification throughout the medium wave band.

This conversion can be carried out in approximately 30 minutes.

THE COMPLETE KIT OF PARTS

includes a Type OL10 Hearing Aid (with Crystal microphone) in perfect working order with miniature ear phone and moulded ear insert attached: ferrite rod, germanium diode, components, circuit diagram and full instructions. Price £2 6s. 0d. post paid (or without crystal microphone £2 2s.).

Batteries extra: 1.5v. L.T. (Type D18), 8d.; 30v. H.T. (Type B119), 4/3.

NOTE: As the crystal microphone is not used in the Pocket Radio, it can, if desired, be used as a general microphone and it does not require a matching transformer.



SEE and HEAR this Miniature POCKET RADIO demonstrated.

DROPPERS. Chassis type heat dissipating droppers for midget receivers. 5000Ω 10W. 3/6 post paid.

MAIN DROPPERS. 7500Ω 50W. 2/6 post paid.

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ABSORPTION WAVE METER. Easily converted to 2 metres or 70 cm. In Copper-plated metal case 3½in. × 4½in. × 5½in. with dial calibrated 0-100 and 80V Neon Tube. Coverage approx. 190-210 Mc/s. New. 6/6 each post paid.

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A superb 8 valve 4 band receiver covering 150-400 kc/s., 400-1100 kc/s., 2-5 Mc/s., and 5-10 Mc/s. Valve line up 6SK7 R/F, 6K8 F/C, Two 6SK7 IF Amplifiers, 6R7 Second Det. AVC and AF Amplifier, 6C5 BFO, 6K6 OP, 6H6 Sig. limiter diode.

SENSITIVITY. 4 mV. signal, gives 50 mW. audio at 4-1 signal/noise ratio. SELECTIVITY. At 5 Mc/s. is 39 kc/s. at 1,000 times down. At 150 kc/s, bandwidth is only 22 kc/s. IMAGE RATIO. Minimum image ratio is better than 1500-1 on 5-10 Mc/s Band and 10,000-1 on 150-400 kc/s.

As a BOAT, TRUCK, CARAVAN or CAR RECEIVER it is UNEQUALLED in value; converted to a.c. operation for fixed station, it equals receivers selling for over five times the price we ask.

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This testmeter has a basic movement of 400 microamps and is calibrated for use on the following ranges:-

A.C. and D.C. Volts 0 to 5,000V in 6 switched ranges.

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This instrument is contained in a well finished polished wood case with leather carry-Leads and ing handle. test probes are housed in the case which measures 6½in. × $6\frac{1}{2}$ in. \times $4\frac{1}{2}$ in.

All meters fully tested before despatch. Supplied complete with moulded test probes, full operating instructions and circuit diagram. PRICE

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3 WAVEBANDS:-L.W. 800m-2000m, M.W. 200m-550m, S.W. 16m-50m Chassis size lâjin. × 5½in. × 2½in. Attractive Glass Dial lôin. × 4½in. edge lit by 2 pilot lamps. Horizontal or Vertical Station Names and 4 control knobs, wainnt or ivory to choice. 4 position W/G switch, L.M.S. and Gram. P.U. sockets. hödern circuitry, all coils asquistable dust cored and only quality components used throughout. Delayed A.V.C. and neg. feed-hack. A.G. mains 200/250 v. Double wound transf. isolates chassis from mains. Aligned and callbrated ready for use.

BRAND NEW & GUARANTEED £9.15.0 Carr. and ins. 4/6. 3-ohm speakers suitable for this chassis available 8" 17/6 10" 25'-

This chassis is a genuine bargain and delivery is reasonably good.

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Kit of parts to build this modern and highly successful unit complete with drilled chassis and J.B. dial, wound coils and screening cans, 4 B.V.A. miniature valves, and all necessary quality components etc. for only 26/10s, post free. Superior dial calibrated me/s., edge lit by 2 pilot lamps, 12/16 extra. Power Pack components kit including double wound Mains Transformer 26/25/- extra.

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TUBULAR	CAN TYPES	80 CABLE CO-AXIAL
25/25 v., 50/12 v. 1/9	8+8/450 v 4/6	SPECIAL. Semi-air spaced polythene,
50/50 v. 4/500v. 2/-	8+16/450 v 5/-	standard lin. diam. Stranded core.
8/450 v 2/3	8+16/500 v 5/8	Losses cut 50% 9d. yd.
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16/450 v 3/6	32+32/450 v 6/6	OUTLET BOXES 4/6
16+16/450 v 5/6	60+250 v 6/8	BALANCED TWIN FEEDER per yd.
32/350 ▼ 4/-	60+100/350 v. 11/6	(80 ohms) 6d. TWIN SCREENED FEEDER per yd.
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32+32/500 v 7/6	1000 +1000/6 v. 6/6	dia.
CONDENSERS Mica,		OITTO 1% (ex stock).
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6d. ea., 5/- doz. Hunts and T.C.C000		I., 2/ PEAKER FRET.—Expanded Bronze
.02 and .1 350 v., 9		nodised metal Sin. × Sin., 2 3; 12in. ×
Hunts Moldseal, 1,		In., 3/-; 12in. × 12in., 4/3; 12in. ×
.5 Hunts, 1/9.	1	6in., 6/-; 24in. × 12in., 8/6.
SILVER MICA CON		YGAN FRET (Murphy pattern)12lu. ×
5 pt. to 500 pt., 1/ 8		2in., 2/-; 12in. × 18in., 3/-; 12in. × 4in., 4/-; etc., etc.
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RAND 3

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Kit of parts complete to build this most successful unit comprising deilled chassis, valves, wound coils, and all necessary components, slightly modified version using 8D3 instead of EF80 valves, £2/5/-, post free.

Provision has been allowed on chassis for small power pack and Aerial switching.

Power Pack Components in Mains Transformer 30/- extra.

TRIMMERS, Ceramic, 4 pt.—70 pt., 9d 100 pt. 150 pt., 1/3; 250 pt., 1/6; 400 pt., 1/9. PHILIPS Beehive Type—2 to 8 pt. or 3 to 30 pt. 1/3 each.

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18 g. Plain undfilled. Folded 4 sides, riveted corners, lattice fixing holes. Depth 2 fin., 7in. × 4in., 4/6; 9in., × 6in., 5/9; 1iin. × 7in., 6/9; 13in. × 9in., 8/6; 14in. × 11in., 10/6, etc.

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K3/25 2 kV., 4/3; K3/40 3.2 kV., 6/-; K3/45 3.6 kV., 6/6; K3/50 4 kV., 7/3; K3/100 8 kV., 12 6; K3/150 1 kV., 7/3; K3/100 8 kV., 12 6; K3/150 1 kV., 16/-; RM1 25 v. 60 mA., 4/-; RM2 L5 v. 100 mA., 4/3; RM3 15 v. 120 mA., 5/104 250 v. 275 mA., 16/-; HT59 250 v. 220 mA., 26/6.

MAINS TRANSFORMERS. Soundmaster, 35/-; Mullard Amp. F/shrouded, 35/-; Ostam 912 F/shrouded 35/-.

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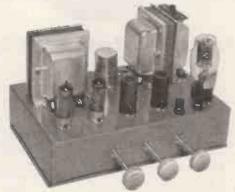
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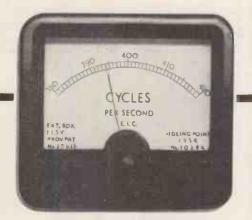
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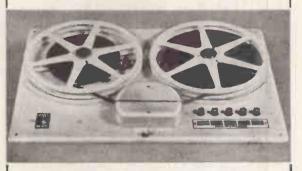
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(ii) CHEMISTRY, bio-chemistry and metallurgy. CHEMISTRY, bio-chemistry and metallurgy.

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mission, Scientific Branch, 30, Old Burlington St., London, W.I. quoting No. S 59/55. [4843]

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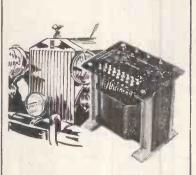
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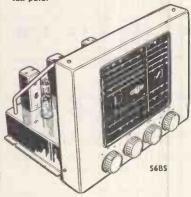
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and quote M2C/40089/WF. [4909]

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experience experience and to the Personnel Manager, The General Electric Co., Ltd., Brown's Lane, Allesley, Coventry. Ref. RG/B. [4710]

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[4883]

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SALES.—Electrical Relay Manufacturers, members of large industrial group, wish to engage a really competent man to handle ton level sales activities.—Write in first instance, stating full qualifications, to Managing Director, Besson & Robinson, Ltd., Harlow, Essex.

ELECTRONIC Laboratory Assistant required measurements; superamulation scheme, required measurements; superamulation scheme, which is and salary required to —Personnel Manager, The and salary required to —Personnel Manager, The The Telegraph Condenser Co., Ltd., North Action, W. Manager, Especial Condenser Co., Ltd., North Action, W. Manager, Carbon Legisland, Ltd., North Action, W. Manager, Carbon Legisland, Ltd., North Action, W. Manager, Carbon Ltd., North Action, W. Manager, Carbon Ltd., North Action, W. Manager, Carbon Ltd., North Action, W. Manager, Ltd., North Action, W. Manager, Carbon Ltd., North Action, W. Manager, Carbon Ltd., North Action, W. Manager, Carbon Ltd., North Action, W. Manager, Ltd., The Telegraph Condenser Co., Ltd., North Acton W.3.

LECTRONICS: technical author for instruction books on electronics and radar subjects; sound practical and theoretical knowleds, sound practical and theoretical knowleds, and ability to write good English essential; author expedence preferred, but engineers with suitable literary ability considered; London area.—Box 4589.

A. H. HUNT (Capacitors). Ltd., require and electrical experience. for work on special purpose machines. Write giving age and previous experience to Personnel Manager, A. H. Hunt (Capacitors). Ltd., Bendon Valley. Garratt Lane. Wandsworth, S.W.18. [0057]

Hunt (Capacitors). Lid., Eendon Valley, Garratt Lone, Wandsworth, S.W.18.

TECHNICAL Publications.—Electronic Devalupment Engineers, Senior and Junior, with a leaning towards Technical Writing to undertake the preparation of Instruction Manuals for Radar and like projects. A working knowledge of Pulse Circuitry and Microwave Techniques essential.

ENIOR Illustrator required, with experience of preparation of circuits, block diagrams, perspective and exploded views.—Apply, giving full details of experience, age, etc., to the Personnel Manager. The General Electric Co., Ltd., Brown's Lane, Allesley, Coventry.

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PYE, Ltd., Radio Works, Cambridge, invites applications for positions as technical authors and illustrators for the compliance of maintenance handbooks dealing with radio equipment: applicants must possess command of English and some technical knowledge.—Applications in writing to Assistant Personnel Officer.

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Surrey. Duties include repair, calibration and modification of electronic apparatus used in R.A.F. hospitals.

AGE at le_st 25 on 1st July, 1955. Candidates must have served an apprenticeship, or have had industrial experience or Service training, in the electronic field; they must also have Intermediate Certificate of the City and Guilds Institute in Telecommunications Engineering or equivalent, or produce evidence of equivalent standard of technical education. STARTING salary £550 (women £501) at age 25 to £580 (women £564) at age 28 or over; maximum £680 (women £611). Women's scales being improved under equal pay scheme. PARTICL ARS and application forms from FARTICL ARS and application forms from FARTICL ARS and application forms from Character (30, 0) of Burlington St., Lundon, W.1. quotting \$484/55. Completed application forms must be returned by 11th August, 1955. [4963]

DESIGNER draughtsman required for radio perienced in surface lay-out and mechanical design for commercial receivers, single man preferred; free air passage, salary £850 or in accordance with ability and experience.—Please reply to Box 4926.

reply to Box 4926.

TECHNICAL authors, senior and junior required for the preparation of maintenance quired for the preparation of maintenance and the preparation of maintenance and the preparation of maintenance and the preparation of the p

Andrew's Road, Cambridge. [4940]

A SSISTANT engineers required for radio communication development work, good prospects for men with H.N.C. or equivalent commencing salary according to age and experience.—Apply in writing, stating age, experience, etc., to Siemens Brothers & Co.. Ltd., Ref. 744/20, Woolwich, S.E.18. [4945]

INSPECTOR required to start new department for test and inspection of electronic equipment; adequate technical knowledge and ability to work with minimum supervision; excellent prospects for advancement in expanding company.—Rivilin Instruments. Ltd., 7a. Maitland Park Villas, London, N.W.5. Gul. 2960, [4970]

MECHANICAL development engineer, with qualifications to at least ONG (Mech.) standard, required for work on radio and television receiver details; previous experience in light mechanical engineering desirable.—Apply. Chief Engineer. Pye. Ltd.. St. Andrew's Rd.. Cambridge.

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A IRCRAFT radio aerials.—Gloster Aircraft Co., Ltd., Gloucester, have a vacancy for an electronic engineer (HNC. or equivalent) to work on the development of suppressed aerials for aircraft; the work involves theoretical and practical knowledge of measurements from "HF." to "X" bands and of the properties of different types of suppressed aerials. A SMALL laboratory is being established for the work: an elementary knowledge of aircraft design and construction would be an advantage.—Applications, stating age, pravious employers and experience, etc., should be addressed to the Employment Officer. [4553]



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Lane, Wandsworth, S.W.18.

SALES engineer required by electrical components manufacturer; these components are being used by leading radio and television manufacturers, and experience in this field is desirable; H.N.C. or equivalent an advantage: the successful applicant will be based on London West End office, current driving licence necessary.—Apply Box 4838. [4923]

necessary.—Apply Box 4838. [4923]

E LECTRONIC engineer required for strain gauge department to assist in maintenance and development of equipment for the measurement of vibration and stress in helicopters and fixed-wing aircraft; applicants should be of degree standard and should write to the Personnel Manager, The Fairey Aviation Co.. Ltd., Hayes, Middlesex, under reference R.D.T.

Ltd., Hayes, Middlesex, under reference R.D.T.

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has a number of vacancies for electrical
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some previous experience of, alternatively, a
real interest in electronics would be an advantage.

tage.

COMMENCING salary will be according to individual qualifications and experience and will provide progressive remuneration for increased responsibilities. Applications in writing, which will be treated in confidence, should be addressed to Mr. J. F. Gogarty, Personnel Officer, The Mullard Radio Valve Co., Ltd., New Rd., Mitcham Junction, Surrey, quoting reference E.1.

ence H.1.

SENIOR Electronic Engineer required for the development of a new project ancillary to the commercial use of large-scale electronic computers in the LEO project of J. Lyons & Co., Ltd. The starting salary will be in the range of £800 to £1,000 according to qualification and experience.—Write, giving full details. to Control Office. Cadby Hall, London, W.14, [4827]

YOUNG men required for assembly, wiring and testing of electronic aparatus, some experience with scientific instruments an advantage, technical school education, must have completed National Service; pension scheme 5-day week, canteen; Camden Town district; salary according to age and experience, —Please apply, stating age and experience, to Box 3389, [4635]

DEVELOPMENT engineers; a leading manu-equipment invites applications from qualified men with experience in this field for respon-sible positions in the laboratories.—Those interested should write, giving brief details of qualifications and experience, to the Personnel Manager (Ref. D.E.), Box 4594.

qualifications and experience, to the Personnei Manager (Ref. D.E.), Box 4594. [4872]
TECHNICAL Sales Assistant with experience, preferably gained in the wholesale trade or as manufacturers' representative, of marketing radios, televisions, cinema equipment and allied items, is required by an old-established British company in Bangkok; good education and appearance, age 23-27.—Write, with full particulars of previous experience, to Box JN/132, c/o 95. Bishopsgate, London, E.C.2. [4933]
ELECTRONIC/electrical engineer required by London manufacturers of radio components to take charge of measurements laboratory and test gear design, experience of measurements either capacitors, inductors or telecommunication cables an advantage; salary range 2750 to £300; company rapidly expanding—write details of experience, qualifications, age and present salary Box 4568. [486]

and present salary Box 4568. [4861]

MAINTENANCE technician required by the National Film Board of Canada for duty in sound recording studios in Ottawa and Montreal. Applicant must have had several years' experience in either Film Broadcasting or Recording studios and should be familiar with transmission testing of audio equipment. Salary \$5.500 by yearly contract with air transportation paid.—Box 4998.

tion paid.—Box 4998. [4978]

A VO—The manufacturers of Avo electrical and electronic instruments invite applications from young men having completed National Service, for a post of Assistant Technical Sales Engineer. Applicants must have Ordinary National Certificate, City and Guilds Radio II, or similar qualification, and will, in addition to other duties, be expected to take a special interest in data relating to receiving valves.—Applications to Mr. R. E. Hill, Acweeco, Ltd., Westminster. London, S.W.1. or Tel. Victoria 3404

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REPLY to Ref. 246 Pye, Ltd., Personnel Dept.. (2496)

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STEWARTS AND LLOYDS, Ltd., have vacancies in their iron and steel works at Corby, Northamptonshire, for maintenance instrument mechanics, where the town, with modern housing, is rapidly expanding and prospects of obtaining a house are good; full instrument apprenticeship or five years' experience maintenance electrical or mechanical instruments essential.—Apply to Labour Service Officer. Stewarts and Lloyds, Ltd., Corby. [435]

Stewarts and Lloyds, Ltd., Corby. [4835]

RADAR and Telecomms.—Service engineers required by British firm in Aden. Applicants should have at least 5 years' experience in marine radio and radar servicing and preferably some V.H.F. experience. Ex-technical officers and senior N.C.O.s from H.M. Forces are particularly invited to apply. Free accommodation, medical attention and first-class passages for wife and family. Commencing salary is dependent upon qualifications and experience but would not be less than £1,000 per annum. Pension scheme.—Apply in writing, stating nationality, personal details, education and experience, to Box 4473. [4841]

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Please write stating age, qualifications and experience to Dept. C.P.S., 336/7 Strand, W.C.2,

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ducts for the electronics, radio and T.V. industries. Physicists and Electronic Engineers are required for development, applications and instrumentation work on semiconductors, having learnings magnetic materials or electrical ceramics, magnetic materials or

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Write in confidence to the Technical Manager, The Plessey Co. Ltd., Towcester, Northants.

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THE ENGLISH ELECTRIC CO., LTD.,

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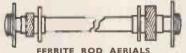
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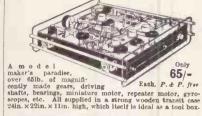
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[4850]

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[4855]

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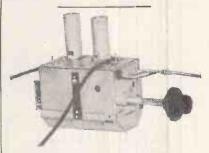
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[14855]

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Egen Electric, Ltd. 56 E.K.E. 162 Electrical Instrument Repair Service, The 162 Electrical Instruments (H'ton), Ltd. 145 Electrica Acquistic Developments 163	Northern Radio Services 25 Nu-Swift, Ltd. 162	West-End Radio, Ltd. 154 Westinghouse Brake & Signal Co., Ltd. 68 Weymouth Radio Mfg. Co., Ltd., The 80 Whysfeddle Wireless Works 58, 111 134
Electronic Precision Equipment 104, 105, 106 Ellison Transformers, Ltd. 110 E.M.I. Institutes 51, 94, 97, 102 English Electric Valve Co., Ltd. 15 Enthoven Solders, Ltd. 22 E.W.A. 138 Excel Sound Services 136	Painton & Co., Ltd. 35 Parker, A. B. 136 Parsonage, W. F., & Co., Ltd. 164 Partridge Transformers, Ltd. 151 P.C.A. Radio 81 Pearce, T. W. 159 Philips Electrical, Ltd. 70	Wilkinsont
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