# Wireless Wprlil 

## Radio Electronics • Television




## Wiirelless World

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## 32. A HIGH GAIN VIDEO AMPLIFIER

The gain of a video amplifier with a given bandwidth, using a high slope r.f. pentode, is limited by the output or load capacitance. A typical cathode compensated circuit using an EF80, with a bandwidth of $3 \mathrm{Mc} / \mathrm{s}(6 \mathrm{~dB})$, a maximum video output voltage of 50 V , and a load capacitance of 26 p F , has a gain of 11 . The gain can be increased and the bandwidth maintained if the effective load capacitance is reduced by means of a cathode follower. The practical circuit shown in Fig. 1 has a gain of 22. A Mullard PCF80 triode pentode combines the functions of video amplifier and cathode follower in one envelope.


## Video Amplifier

The cathode of the pentode (Fig. 1) is held at about +3.5 V by the pentode cathode current and the bleed current through the $27 \mathrm{k} \Omega$ resistor. With the screen grid at 190V this gives an anode current of 3.5 mA . The permissible anode current swing is limited by the maximum screen grid dissipation rating of 0.5 W , therefore the output voltage is limited to about 80 V . The dynamic characteristic of a typical PCF80 pentode shows that the corresponding maximum limit of the anode current is 8.8 mA . The effective slope is $1.5 \mathrm{~mA} / \mathrm{V}$, therefore the gain $\mathrm{gm}_{\mathrm{m}} \mathbf{R}_{1}=$ $1.5 \mathrm{~mA} / \mathrm{V} \times 15 \mathrm{k} \Omega=22.5$.

## Cathode Follower

The total capacitive load of a video stage is made up, in a typical good design, approximately as follows: Video output 4 pF , C.R.T. 8 pF , Synchronising separator 6 pF , Noise suppressor 2 pF , Strays $6 \mathrm{pF}=26 \mathrm{pF}$.

If the total gain of the stage is to be greater than 20 and a cathode follower output (with a gain less than 1) is to be used, the amplifier must have a higher gain-say 25 . To achieve this the output capacitance must be less than 11 pF . With a cathode follower the capacitive load of the amplifier is reduced to: Video output 4 pF , Cathode follower input 3 pF , Strays $3 \mathrm{pF}=10 \mathrm{pF}$, which is within the 11 pF limit. The overall response is limited by the cathode follower, which must be capable of developing an output voltage of at least 50 V across a capacitive load of 26 pF .

Further notes on the circuit are available with reprints of this advertisement. Reprints of all advertisements in the Valves, Tubes and Circuits series may be obtained free from the address below.

## Cathode Current Requirements

The cathode current has two components: a current which maintains the signal across the cathode resistor, and a transient current which charges the output capacitance. The valve must supply both currents. If it failed to do so the cathode voltage would not follow the grid voltage. If the cathode cannot rise at the same rate as the input the valve will run into grid current, and the positive going edge of the signal will be distorted. Alternatively, if the cathode cannot fall at the same rate as the input, the valve will be cut off and the negative going edge will be distorted. In general the second alternative is worse, as cut-off is an absolute limit to change in the cathode current. The onset of grid current does not preclude further change if the source impedance of the driver stage is reasonably low. As each input pulse has two transients associated with it the total current swing is the maintenance current plus twice the transient current. To prevent cut-off the standing cathode current must exceed the sum of the transient and maintenance currents.

FIG. 2


## Derivative Correction

It is possible to compensate for a poor fall time in a cathode follower by adding to the output a voltage proportional to the derivative of the video input signal. This voltage, which can be obtained by the inclusion of a resistor in the triode anode circuit, may be a.c. coupled to the c.r.t. grid when the cathode follower is driving the c.r.t. cathode. The derivative correcting voltage is thus added in the correct phase. It is found that this arrangement increases the contrast of the high definition bars of test card ' C ' and improves the vertical edges; but it tends to produce ringing and $2.5 \mathrm{Mc} / \mathrm{s}$ oscillation, therefore the use of derivative correction with this circuit is not recommended.

## Performance

In a 21 -inch tube receiver the recommended circuit (Fig. 1) gives a composite video output up to 80 V before there is any visible sign of cathode follower overload. The frequency response, measured with a sine wave input and a 50 V peak-to-peak output, is shown in Fig. 2. The stage gain is 22 .

## Wirplase Sorflil

AUGUST 1955
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## Tape Bookimarfik

THE rapid growth of tape recording for domestic and business purposes is largely duc to the excellence of what is rather pompously called the tape transporting system. Mechanisms for winding the tape from one spool to the other are, generally speaking, remarkably effective; the clumsiest of us find little difficulty in operating even the cheaper and less refined machines.
Judging by letters from readers, however, the domestic tape recorder is widely thought to be deficient in one respect: it is difficult to "find the place" on the tape at which any particular item starts. True, some kind of rough-and-ready gauge is often provided, but it does not give precise location. Most home recordists make use of scraps of paper inserted, like a bookmark, between adjacent turns of tape. This method, though fairly adequate when the "fast wind" mechanism can be smoothly and accurately controlled, is often considered to be crude. What seems to be wanted by a wide circle of users is a device which, by mechanical, magnetic, electrical, or even electronic means, will indicate precisely the part of the tape required. Something much more precise than the classical bookmark is in fact called for.

A device that went a long way towards meeting these needs was described in Wireless World for April, 1955. This embodied a selector switch actuated by pads glued to the tape at appropriate positions. Another method, used in certain machines, calls for the use of a revolution counter.

Perhaps the most novel of the various place-finding suggestions put forward by readers is for a system of signals imposed on the tape and audible only during the "fast wind"-forward or back. The idea is that these pulse signals, which could be arranged according te a code, would be recorded at a very low frequency and so would not be heard (or at least would not be obtrusive) at normal playback running speed. This method has obvious attractions-and, probably, what might be considered equally obvious limitations and drawbacks. For one thing, it would not be readily applicable to existing machines in which the "playback head is inoperative during "fast wind." However, the
method seems worth investigating-always assuming, of course, that no satisfactory alternative that is inherently simpler and cheaper can be devised.

## Cofonr Television <br> Exporignouls

TOO much significance need not be attached to the standards (set out on another page) chosen for the B.B.C. experimental transmissions in colour television, to begin in the autumn. It was more or less a foregone conclusion that a start would be made with a version based as closely as possible on the standards devised by the National Television Standards Committee of the U.S.A. The so-called "British N.T.S.C." system involves very little change or addition to the B.B.C.'s existing monochrome transmitters. Another factor which might well have influenced the decision is that there is now a large amount of American experience with this system on which to draw. Further, there can be no doubt that the technical elegance of the system is attractive, except perhaps for the complexity of the receivers that go with it. Presumably the main purpose of the experimental transmissions is to allow that section of industry concerned with receiver design and manufacture to carry out field tests and to gain practical experience of colour television.

So far there has been no mention of experimental transmissions using other systems. No doubt, however, these will come; there is no need to make an over-hasty decision as to what system is finally to be adopted in this country. Indeed, a regular colour service is not expected for a long time.

In considering these questions, it must be remembered that all factors affecting British television must finally be decided by the Postmaster-General, who, in his turn, is advised by the Television Advisory Committee. The T.A.C. has uncompromisingly recommended a compatible system, and the British N.T.S.C. standards of the forthcoming transmissions satisfy that requirement. However, there may well be a change of view on this subject before the time comes for making a final decision.

## Groucth of Aircraft Radio

IN the Golden Jubilee celebrations of the Royal Aircraft Establishment at Farnborough this year, a substantial proportion of the exhibits were concerned with radio aids to navigation, communication and the application of radio and electronic techniques to the guidance of rocket missiles and the simulation and computation of their performance.
An historical exhibit, arranged in chronological order, began with the Sterling Type 52A spark transmitter and the Model TF valve receiver (detector and 2 l.f. stages) of the 1914-18 war, and led up to the multi-channel crystal-controlled equipment in use to-day. One was reminded of the fact that although the possibilities of flying and of wireless communication were both realized at the turn of the century, a decade was to pass before they could come together. With the aerofoils and engine powers then available no designer, harassed as he was by thoughts of how best to save weight, could afford a second glance at the ship-and-shore type of equipment which was the stock-in-trade of the then infant radio industry.
There is still constant pressure from commercial and military aireraft designers to reduce the weight and size of radio equipment to make room for paying passengers or more armament, and this is being met by increasing miniaturization, which was adequately represented in the exhibition.

The current work of the Radio Department at R.A.E. covers a wide field including the development of sono-buoys (in which the noises of submarines are picked up by hydrophones and relayed to a searching
aircraft by radio) and the detailed investigation of the problems of installing aerials with reasonably omni-directional characteristics on high-speed aircraft.

Electronic methods, once a useful alternative to established methods of physical measurement, can now be said to dominate all branches of aeronautical research. They reach their zenith at Farnborough in the "Tridac" analogue computor for guided missile problems, which occupies the whole of a special building and calls for primary power of the order of hundreds of kilowatts for the functioning of its many circuit elements.

## Disc Recording and

## Reproducing Characteristics

IT has often been said that the ideal disc recording or reproducing characteristic should be one which is also easily realizable with simple circuitry. This provision is met in the proposals contained in the revised Eritish Standard 1928:1955 which gives curves for fine-groove and coarse-groove recordings, and formulæ for their derivation in terms of the time-constants of simple R-C networks. These recommendations are based on C.C.I.R. standards and take into account the recommendations provisionally agreed by the International Electrotechnical Commission at their Philadelphia meeting last year.

The revised standard, which is obtainable, price 6 s , from the British Standards Institution, 2, Park Street, London, W.1, includes specifications of commercial and transcription disc dimensions, stylus tip radii and concludes with a discussion of the arguments for standardization of the recording and/or the reproducing characteristic.


## Radar Simulator

EQUIPMENT for the training of radar operators and for the synthesis of tactical air exercises without the use of aircraft has been designed and developed by C. E. G. Bailey and J. Somerset Murray in association with the Solatron Electronic Group, Ltd., Thames Ditton, Surrey.
Aircraft are represented by control units which feed signals (analogous to speed, rate of climb or dive, rate of turn and direction of flight) into a computor unit. This unit integrates the factors governing range, bearing and height and translates the result into pulses for transmission to one or more radar display units.
Extraneous effects such as tropospheric refraction can be taken into account, and there are facilities for simulating jamming, either of the reflecting type ("window") or of the active noise-generating type.
The essence of the design is flexibility so that future as well as current characteristics of aircraft performance and radar systems can be simulated.

Typical assembly of Solatron radar simulator units. The main cabinet houses the computors. A display unit is shown on top. In the foreground are two aircraft control units.

# cololr television standards 

For Forthcoming B.B.C.

Test Transmissions

E$\triangle$ XPERIMENTAL colour television transmissions based on the American N.T.S.C. compatible system will be made by the B.B.C. from Alexandra Palace this autumn as part of the general investigations into the best type of colour system for this country. A specification of the standards to be used shows the method of transmission to be a scaled-down version of the N.T.S.C. system, with the colour information transmitted by means of a sub-carrier within the existing $6.75-\mathrm{Mc} / \mathrm{s}$ monochrome channel.

An article on page 393 of this issue elaborates on the principles of the " British N.T.S.C." system, so it is unnecessary to add here more than the bare facts of the B.B.C. specification. First of all, the existing black-and-white transmission from Alexandra Palace will remain as it is and form the "luminance" or brightness component of the complete colour signal. Simultancously a "chrominance" or colouring signal will be transmitted in the form of two sets of a.m. sidebands of two suppressed carriers in quadrature, these having the common frequency of $525 / 2$ times the line scanning frequency relative to the $45-\mathrm{Mc} / \mathrm{s}$ picture carrier in fact $2.6578125 \mathrm{Mc} / \mathrm{s}$. These two "chrominance" components, known as the $\mathrm{E}_{\mathrm{I}}$ and $\mathrm{E}_{Q}$ signals (see article) will carry respectively wideband colour information up to $1 \mathrm{Mc} / \mathrm{s}$ and narrowband colour information up to $340 \mathrm{kc} / \mathrm{s}$.

The "chrominance" or colouring sync signal will consist of a reference burst of 9 cycles of sub-carrier frequency transmitted during the "back porch" black-level period following each line sync pulse. It will not occur during the eight broad pulses of the frame sync period. Details can be seen from Fig. 1, which shows that the sync burst penetrates into the picture region and that the "chrominance" signal can do the same and also rise above peak white.

The complete colour signal, in terms of the total


Fig. I. Video waveform showing the relation of the added chrominance signal and sync burst to the existing monochrome signal.
video voltage applied to the transmitter modulator, is composed as follows:-
$\mathrm{E}_{\mathrm{M}}=\mathrm{E}_{\mathrm{Y}}+\mathrm{K}\left\{\mathrm{E}_{\mathrm{Q}} \sin \left(\omega \mathrm{t}+33^{\circ}\right)+\mathrm{E}_{\mathrm{I}} \cos \left(\omega \mathrm{t}+33^{\circ}\right)\right\}$
Here $\mathrm{E}_{\mathrm{Y}}$ (the "luminance" or black-and-white signal) is made up of $0.3 \mathrm{E}_{\mathrm{R}}+0.59 \mathrm{E}_{\mathrm{G}}+0.11 \mathrm{E}_{\mathrm{B}}$ while $\mathrm{E}_{\mathrm{Q}}=$ $0.41\left(\mathrm{E}_{\mathrm{B}}-\mathrm{E}_{\mathrm{Y}}\right)+0.48\left(\mathrm{E}_{\mathrm{R}}-\mathrm{E}_{\mathrm{Y}}\right)$ and $\mathrm{E}_{\mathrm{I}}=-0.27\left(\mathrm{E}_{\mathrm{B}}\right.$ $\left.-\mathrm{E}_{\mathrm{Y}}\right)+0.74\left(\mathrm{E}_{\mathrm{F}}-\mathrm{E}_{\mathrm{Y}}\right)$. The angular frequency $\omega$ is $2 \pi$ times the frequency of the "chrominance" subcarrier, while the phase reference is the phase of the sync burst plus $180^{\circ}$. The factor K indicates that various ratios of "chrominance" to "luminance" between 1.0 and 0.3 may be used in certain experiments.

Comparative tests may be carried out with the system locked to and unlocked from the $50-\mathrm{c} / \mathrm{s}$ mains. With unlocked operation the sub-carrier frequency will be the $2.6578125 \mathrm{Mc} / \mathrm{s}$ mentioned above, but in the locked condition it, and the frequency difference between the vision and sound carriers, will change directly with the mains frequency.

The experimental transmissions will, of course, be made outside normal programme hours, and readers who manage to receive them will perhaps get a foretaste of the effect of compatible colour-detrimental or otherwise-on the picture quality we are used to at present.

SHORT-WAVE CONDITIONS Predictions for August


THE full-line curves given here indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during August.
Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.

frequency below which communication should be possible for $25 \%$ of the total time

-     - predigted average maximum usable frequency
—— frequency below which communication should
be possible on all undisturbed days


# WORLD OF WIRELESS 

Show News * Anglo-French TV Link *<br>New Television Stations

## Earls Court

NEARLY 50 per cent of the 90 or so manufacturers who will be exhibiting at this year's National Radio Show are set makers; the others being makers of components and accessories. The remaining 30 exhibitors are either users of radio-such as the B.B.C., the Services and Government Departments-wholesalers, societies and associations, and organizations providing services for the industry.

The show opens at Earls Court, London, S.W.5, on August 24 th for ten days, with a preview for overseas and invited guests on the 23 rd .

In addition to the facilities provided on the exhibitors' stands for the demonstration of Bands I and III television sets, there will again be a display of some 100 receivers in Television Avenue. These receivers will be tuned to Band I. Incidentally, as in previous years, the Band I carriers piped round the show will be in Channel 4 to avoid interference from the London transmitter. Channel 8 is being used for the Band III demonstrations.
Each day at 2.30 - except on Saturdays-a discussion meeting is being arranged by the British Radio Equipment Manufacturers' Association to bring together retailers' servicemen and representatives of the industry to discuss the servicing of f.m. receivers and the problems of f.m. aerial installation. Tickets must be obtained from B.R.E.M.A., 59, Russell Square, London, W.C. 1 .

## Cross-Channel TV Link

THE first section of the permanent Anglo-French television link, ordered by the B.B.C. from the Post Office last January, will be completed in September. This section consists of a two-tube co-axial cable between London and St. Margaret's Bay, Kent. The next section, consisting of a two-way cross-Channel radio link, will not be completed for three years.

However, so that we can participate in international programme exchanges (the next is planned for this autumn) the cables will be extended temporarily from St. Margaret's Bay to Swingate, near Dover-a distance of about two miles-where a temporary crossChannel radio station is set up.

It is understood that the cost of laying the cable, which is permanently rented by the B.B.C., and the cost of building the terminal station at St. Margaret's Bay will be nearly $£ 750,000$.

## NATIONAL RADIO EXHIBITION WIRELESS WORLD SHOW NUMBERS

September: Show Guide. Plan of the stands at Earls Court, with stand-to-stand guide to the exhibits.

[^0]TEMPORARY 200-ft mast corrying the 8-stack orray for the Croydon I.T.A. transmitter. Marconi's are supplying the complete installation for the temporary station due to open on September 22nd. The mast was completed early in July and the sound transmitter is installed.


## B.B.C. Television Progress

DURING the past few days the permanent television station at Divis, near Belfast, has been brought into service by the B.B.C. It replaces the temporary mobile station which has been in use at Glencairn for the past two years. The transmitter, which has an e.r.p. of 20 kW , shares Channel 1 (vision $45 \mathrm{Mc} / \mathrm{s}$, sound $41.5 \mathrm{Mc} / \mathrm{s}$ ) with Alexandra Palace.

It will continue to receive its programmes directly by radio from Kirk o' Shotts, Scotiand. As has been done in the past, if the received picture is unsuitable for retransmission, or below standard, a warning signal -a vertical white bar-will be radiated. This warning signal is also being used by the Norwich transmitter which re-radiates London's transmissions.

The construction of the $640-\mathrm{ft}$ self-supporting tower to carry the aerials for the new B.B.C. London station at Crystal Palace is well under way. At the $440-\mathrm{ft}$ level will be installed parabolic aerials for receiving outside broadcasts. The Band I transmitting aerials -consisting of eight stacks of four dipoles-will be mounted on the section between 440 and $600-\mathrm{ft}$, above which there will be a $40-\mathrm{ft}$ topmast. The tower is being erected by B. I. Callender's Construction and the aerial by Marconi's.

## I.T.A. Links

A RADIO link is to be provided by the Post Office between Birmingham and the I.T.A. Midland station near Lichfield. It will provide two channels to Lichfield and one in the reverse direction. The Lancashire station at Winter Hill, near Bolton, will be linked with Birmingham by cable.

All the equipment for the radio link is being supplicd by the G.E.C. who are also supplying the repeaters and terminal equipment for the cable link.

## PERSONALITIES

Dr Robert Cockburn, M.Sc., Ph.D., A.M.I.E.E., principal director of scientific research, guided weapons and electronics (M.o.S.) since March, 1954, has been appointed deputy controller of electronics in the Ministry in succession to Rear Admiral G. Burghard, whose tour of duty has expired. Dr. Cockburn, who is 44, was scientific adviser to the Air Ministry before joining the Ministry of Supply. For some time during the war he was head of the counter-measures group of T.R.E. and was nead of the counted the American medal of nerit for his work in this field.

Dr. A. L. Cullen, Ph.D., B.Sc., A.M.I.E.E., is to occupy the newly created chair of electrical engineering in the University of Sheffield. After graduating at the Imperial Collcge of Science and Technology, London, in 1940, he joined the staff of the Royal Aircraft Establishment, Farnborough, where, until 1946, he worked on radar, being mainly concerned with acrials and waveguide techniques. He is at present reader in electrical engineering at University College, London, which he joined in 1946 as a lecturer. Dr. Cullen's special interest is microwave measurement techniques.
J. A. Smale, engincer-in-chief of Cable and Wireless since 1948, has been appointed by the government of Cyprus to be part-time chairman of the new Cyprus Inland Telecommunications Authority set up to administer and operate the island's inland telephone and telegraph services. He will continue in his present post with C. \& W., visiting Cyprus as necessary.
H. R. Whitfield, M.I.E.E., who has been with Kelvin and Hughes, Ltd., as chief radar engineer since 1946, has been appointed a director of Kelvin Hughes (Marine), Ltd. In 1936, at the age of 19, he went to the Automatic Telephone and Electric Co., Liverpool, as a transmission laboratory engineer. From 1938 to 1940 he was a member of the War Department civilian technical staff attached to the Bawdsey (Suffolk) radar research station and then went to the rescarch establishments, Malvern, as a member of the scientific staff on gunnery radar.

In addition to those mentioned in our last issue as having received Birthday Honours, John N. Toothill, general manager, Ferranti, Ltd., Edinburgh, was appointed C.B.E., Horace D. McD. Ellis, engineer in the B.B.C. Designs Department, was appointed M.B.E. and William Fairhurst, foreman, Electronic Tubes, Ltd., High Wycombe, received the British Empire Medal.
R. E. Burnett, M.A., A.M.I.E.E., A.Inst.P., has relinquished the principalship of Marconi College, Chelmsford, which he has held since 1950 and the position of manager of education and technical personnel on his appointment as full-time assistant to the general manager. His duties as manager of education and technical personnel will be undertaken by E. R. L. Lewis, M.A., A.M.I.E.E., who has been his deputy. The new principal of the college is R. G. Hulse, B.Sc., who has been deputy principal during the past year.
L. Hampson, who in this issue describes a Band II tuner unit, has been in the valve measurement and application laboratory of Mullard's since 1951. He joined Mullard's immediately after graduating at the age of 24 at Manchester University where he studied electrical engineering after completing his National Service. While at Mullards he has been mainly concerned with development work on the application of valves in the v.h.f. bands.

After 36 ycars' service with Marconi's, H. C. Van de Velde has relinquished the position of deputy to the managing director of the Marconi Marine Company but will continue to represent the company on the boards of various associated companies overseas. With Mr. Van de Velde's retirement, R. Ferguson, the general manager, who joined the seagoing staff of Marconi's in 1910, is extending his managerial responsibilities. He was seconded to the M.W.T. Company in 1934 in order that he might
take up the appointment of general manager of Egyptian State Broadcasting which Marconi's conducted for the Egyptian government. On his return from Egypt he was re-appointed to the marine company, of which he became general manager in 1947.
F. G. Robb, chicf of Marconi's Test Division since 1948, has retired after 36 ycars' scrvice with the company. He was for some years in the designs and development section where at one time he worked on the development of beam transmitters for the MarconiFranklin short-wave beam system. During the war he was seconded to the Admiralty and became chief of radar

R. FERGUSON

E. H. EVANS
test. His successor at Chelmsford is E. H. Evans who joined Marconi's in 1913. He has been associated with the test division throughout his service and for a number of years has been chief of receiver test.

Obituary. - The death occurred on July 9th of L. F. Fogarty, M.I.E.E., who was for many years honorary treasurer of the Wireless Society of London, of which he was a founder-member and then of the Radio Society of Great Britain as it has been known since 1922. He had been managing director of the Zenith Electric Companv since its formation in 1918.

## IN BRIEF

Receiving Licences Decrease.-Although there was a further increase of 43,192 television licences in the United Kingdom during May, there was an overall decrease of some 17,000 in the number of domestic receiving licences in force. The comparative figures for May and Aprilthe latter in brackets-are: sound 9,102,995 (9,165,242), vision $4,623,917(4,580,725)$, car radio $273,883(271,480)$, total $14,000,795$ ( $14,017,447$ ).
G. A. Briggs, of Wharfedale Wireless Works, Bradford, whose two lecture-demonstrations in the Royal Festival Hall, London, have created a demand for tickets far in excess of the seating capacity of the hall ( 3,000 ), is going to New York to give a similar demonstration in the Carnegie Hall (seating 2,760) on October 9th. Capitol Records Inc. are to make the recordings required for comparing live and recorded performances of piano, violin and organ and, as at the R.F.H., Mr. Briggs will be working in collaboration with P. J. Walker.

At the annual general meeting of the Television Society, the following members were elected to fill the vacancies on the Council: T. W. Price (Ediswan), A. E. Sarson (Marconi's), W. R. Smith (G.P.O.), Professor Trewman (E.M.I. Institutes) and C. B. Townsend (G.E.C.). In addition, D. N. Corfield (S.T.C.) and F. Livingston Hogg (Livingston Laboratorics), who were co-opted last year, were elected full members.
B.A.T.C.-A convention of the British Amateur Television Club is being arranged for October 1 st at the Bedford Corner Hotel, Bedford Square, London, W.C.1,
from 10 a.m. to 6 p.m. There will be a display and demonstration of members' equipment and a film show. Tickets (costing 5s) and further information can be obtained from D. S. Reid, 4, Bishop Road, Chelmsford, Essex.

Telesurance Limited, which operates a television insurance and maintenance scheme through registered R.T.R.A. dealers, has issued a statement on its policy regarding sets converted for Band III. No additional premium charges will be made providing the sets are converted by appointed dealers in accordance with the recommendations of the manufacturers.
The mains and output transformers and the smoothing choke specified by W. A. Ferguson for the 20-watt quality amplifier described in our May and June issues, are being produced by Partridge Transformers, Limited, Roebuck Road, Tolworth, Surrey. A leaflet giving electrical and physical characteristics is available.
In preparation for the advent of commercial television a television training centre has recently been opened in London by Marconi's. It provides a complete training course in the operation and maintenance of television studio equipment. The centre is also available to organizations for rehearsals under operational conditions and a mobile unit is available for the production of recorded O.B.s.

Engineering Education.-Lists of colleges in London and the Home Counties providing engineering courses during the 1955/6 session are given in "Engineering Education in the Region" published by the Regional Advisory Council for Higher Technological Education. It includes sections covering engineering crafts (including radio and television servicing), City and Guilds courses in telecommunications engineering, H.N.C. courses in electrical engineering with a bias towards radio and telecommunications, and courses in direct preparation for I.E.E. and Brit.I.R.E. examinations. The 30 -page booklet is obtainable (price 1s) from the Regional Advisory Council, Tavistock House South, Tavistock Square, London, W.C.I.

At the eighth annual presentation of diplomas at the College of Aeronautics, Cranfield, Bucks, where a chair in aircraft electrical engineering has now been established, nine of the 68 awards were gained by students specializing in aircraft electrical engineering. The theses covered work on computors, simulators, servomechanisms and on the properties of suppressed aircraft aerials. As already announced, the first professor of the new department is G. A. Whitfield, who was at R.A.E. Farnborough.
"The Inquiring Mind" is the title of the documentary film sponsored by the I.E.E. for the purpose of depicting the diverse fields of opportunities open to electrical engineers. Copies of this 30 -minute monochrome sound film, which is available in 35 mm and 16 mm , can be borrowed by schools, colleges and similar establishments. Particulars regarding the loan of the film are obtainable from the I.E.E., Savoy Place, London, W.C.2.

Nuclear Electronics.-During the forthcoming international conference at Geneva on the peaceful uses of atomic energy (August 8th to 20th) to which 84 nations have been invited, two exhibitions are being held. At one of these, which will be open to the public, a number of British electronics manufacturers will be participating and there will be a combined exhibit organized by the Scientific Instrument Manufacturers' Association.
R.E.C.M.F. Headquarters.- The Radio and Electronic Component Manufacturers' Federation has moved from Surrey Street, London, W.C.2, to 21, Tothill Street, London, S.W.1. (Tel.: Abbey 4226/8.)

Swindon.-With the object of forming a radio club in the town a meeting will be held on August 31 st at 7.30 , at the Connaught Café, 34, Cromwell Street, Swindon. Further pasticulars are obtainable from R. Reynolds,

Many demonstrations of high-quality reproduction have been given to gramophone and music societics and at public exhibitions by Goodmans Industries during the past few months. Societies interested in receiving a visit from the demonstration team are invited to write to Goodmans Industries Limited, Axiom Works, Wembley, Middx.

The third edition of the CABMA Register (1955/6) of British products and Canadian distributors again incorporates an alphabetical buyers' guide to some 4,000 British products, including radio and electronic equipment, available on the Canadian market. Other sections list manufacturers, trade names, Canadian distributors, etc. The register (price 2gns) is published jointly by Kelly's Directories and Iliffe and Sons for the Canadian Association of British Manufacturers and Agencies which operates British Trade Centres in Toronto, Vancouver
and Montreal.

## EXHIBITIOV NEWS

Next Physical Society Exhibition.-It is announced by the Physical Society that next year's 40th exhibition of scientific instruments and apparatus will be held in both the Old and New Halls of the Royal Horticultural Society, Westminster, London, S.W.1, from May 14th to 17 th.
Marine Exhibition.-A number of radio manufacturers specializing in marine equipment or industrial electronic gear will be among the 500 exhibitors at the Engineering, Marine and Welding Exhibition which opens at Olympia, London, on September lst for 13 days. Manufacturers appearing in the provisional list of exhibitors include B.T.H., Deca, English Electric (Electronics Division), G.E.C., I.M.R.C., Metropolitan-Vickers, Mullard, Radio Heaters, Redifon and Stratton. The exhibition will open daily from $10 \mathrm{a} . \mathrm{m}$. to $8 \mathrm{p} . \mathrm{m}$.
The dates for the ninth annual Amateur Radio Exhibition, organized by the Radio Society of Great Britain, have now been confirmed; November 23 rd to 26th. It will again be held at the Royal Hotel, Woburn Place, London, W.C.1, and will be opened at noon on the 23rd by Vice-Admiral J. W. S. Dorling, director of the Radio Industry Council.

Midlands Radio Show.-The success of last year's Nottingham Radio Exhibition has prompted the organizers--the Nottingham Centre of the Radio and Television Retailers' Association-to broaden the scope of this year's show and to re-name it the Midlands Radio Exhibition. It will be held in the Ice Stadium, Nottingham, from September 19th-24th.

At the British Exhibition in Copenhagen (September 29 th to October 16th) arranged jointly by the British Import Union, Denmark, and the Federation of British Industries, a large stand has been taken by the Radio and Electronic Component Manufacturers' Federation. Some twenty member-firms will be participating.
The Model Engineer Exhibition, with which is combined this year the Exhibition of Inventions, is to be held at the new Horticultural Hall, Westminster, from August
17th to 27 th.

## BUSIVESS NOTES

Collins Radio, the well-known American manufacturers of aeronautical radio equipment, have formed a subsidiary in this country-Collins Radio Company of England, Limited. At present it is operating a service depot at Sunflex Works, Colham Mill Road, West Drayton, Middx. (Tel.: West Drayton 2226.)

Ekco-Dynatron Merger.--E. K. Cole, Ltd., have acquired a controlling interest in Dynatron Radio, Limited, of Maidenhead, Berks. It is understood that there is no intention of changing the policy of the business which is concerned with the manufacture of "aboveaverage" domestic sound and television receivers.

The potential output of G.E.C. television receivers will be increased by 50 per cent with the rearrangement of production facilities in Coventry. The factory in Spon Street is being devoted exclusivcly to television and the production of domestic sound receivers is being transferred to another factory in the city.

The exclusive world distribution of acoustical equipment developed by Kelly Acoustics Limited, of 295, Regents Park Road, London, N.3, has been taken over by Thermionic Products Limited, Hythe, Southampton, to whom all enquiries for the new ribbon loudspeaker (RLS/1) should be sent.

Ampex Corporation, of California, manufacturers of magnetic-tape recording equipment, are to open an office in Loadon. This and similar offices in overseas countries will be run by the recently formed company, Ampex International.
The "Radiovoyce" microphone equipment illustrated on p. 312 of our July issue, for which Leevers-Rich Equipment Limited are sole marketing agents to the trade, is designed and manufactured by F. W. Hopwood (Developments) Limited, 181, Wollaton Street, Nottingham.

Truvox Limited announce that they have appointed A. B. Thompson (Ireland) Limited, of 15, Newforge Lane, Belfast, as Northern Ireland agents for their taperecording components and accessories.
F. W. Electronics, Limited, of New Southgate, London, N.11, which was formed in 1950 and specializes in the design and manufacture of audio and r.f. equipmentincluding equipment for schools-has moved its works and registered office to 12 a , Prince of Wales Road, Hendon, London, N.W.4. (Tel.: Sunnyhill 0683.)

Holiday and Hemmerdinger, Limited, of 74-78, Hardman Street, Deansgate, Manchester, 3 (Tel.: Deansgate 4121) have notified us that they have arranged with the Recording. Equipment Division of E.M.I. Sales and Service, Limited, to distribute "Emidisc" lacquer recording blanks and accessories to the trade in the Manchester area.

The telephone number of Nagard, Limited, designers and manufacturers of electrical instruments for research and industry, of 18, Avenue Road, Belmont, Surrey, has been changed to Vigilant 9161.

The North-Western Gas Board, which operates two base stations at Manchester and Liverpool and 25 mobile radio-telephone installations, is now using a radio-equipped mobile "paying-in" office. The a.m. equipment being used in the vehicle, which periodically visits outlying districts, was supplied by the Radio \& Transmission Division of Automatic Telephone \& Electric Company. Two additional fixed stations are to be brought into operation by the Board to serve the Wirral and St. Helens areas.

Jack Davis (Relays), Ltd., is moving this month from Percy Street, London, W.1, to Tudor Place, Tottenham Court Road, W.1. Telephone numbers are unchanged.

HALF A MILLION television receivers have come off this assembly line at the Enfield. Middlesex, foctory of Ferguson. The conveyors in the foreground ore carrying completed chassis, ofter soak tests, to the section for final testing and adjustment.

## EXPORT NEWS

Equipment for a public radio-telephone service in the three neighbouring territories of Sarawak, North Borneo and Brunei is being provided by A.T.E. (Bridgnorth), Ltd., a subsidiary of Automatic Telephone \& Electric Company, Ltd. More than 80 single-channel radio links will be required for the service which will link the outlying settlements and the divisional centres where line telephone services already exist. Eventually the main centres will be linked by a multi-channel radio system.

Redifon radio-telephone equipment is being fitted in ten of the vessels of the Niger river fleet of the United Africa Company and fixed stations will be set up at Burutu, Makurdi, Yola and Garua.
A radio network providing for 240 telephone circuits and a two-way $525-$ line television channel between Osaka and Fukuoka has been ordered from Standard Telephones and Cables through its associates the Nippon Electric Company, Limited, of Tokio. Eleven intermediate repeaters, working in the s.h.f. band ( $3,000-30,000 \mathrm{Mc} / \mathrm{s}$ ), will be used to cover the 385 miles.

India's Director General of Supplies and Disposals (Shahjahan Road, New Delhi), has asked for tenders for 12,500 broadcast receivers and associated aerial equipment and loudspeakers. The majority of the receivers are for dry-battery operation and must cover the medium-wave band although some of them must also cover the short waves. About 1,000 receivers are needed for a.c. operation. A copy of the tender documents is available from the Export Services Branch, B.o.T., Lacon House, Theobalds Road, London, W.C. 1 (Reference ESB14288/55). Closing date for tenders is July 29 th .

International Aeradio, Limited, has received from the Egyptian Air Force an order for three air traffic control desks. I.A.L. is also providing the radio and air traffic control services at Yeadon Airport, near Leeds, which is to be developed to provide scheduled and charter air services.

Airfield control radar equipment (Type 424) is to be supplied by Decca Radar, Limited, for installation at Durban National Airport. This surveillance radar equipment, which was reviewed in our November, 1953, issue, was also recently supplied to the South African Air Force.
R. B. Page, of Birmingham Sound Reproducers, Ltd., is on a three-month visit to North America to renew acquaintance with radio-gramophone manufacturers.


# Inexpensive Wave Analyser 

"Zero-beat" System Using Simple Low-Pass Filter

By M. G. SCROGGIE, в.Sc., M.I.E.E.

IN a recent survey of distortion-measuring technique ${ }^{1}$ it was noted that published data on distortion usually take the form of a single figure ("total harmonic distortion") whereas for fair comparison one must know about the individual distortion products (whether harmonics or intermodulation) making up this total. One reason for the scarcity of analysed data is no doubt the high cost of wave analysers. In order to measure each distortion product separately it is necessary to have extremely high selectivity, which cannot reasonably be obtained by straightforward a.f. tuning capable of being varied continuously from, say, $20 \mathrm{c} / \mathrm{s}$ to $20 \mathrm{kc} / \mathrm{s}$. The difficulty is usually overcome in the same way as in the analogous r.f. problem in radio receivers, with the aid of the superheterodyne principle. By means of a beat oscillator, the frequency of the chosen component of the signal being analysed is transferred to an "i.f.," which might be $50 \mathrm{kc} / \mathrm{s}$; components of all other frequencies are then removed by a filter having a pass band of only a few cycles per second, and after amplification the selected signal deflects an indicating meter.


Fig. I. Block diagram of the type of wave analyser described.
The heart of the instrument is the filter, because on it the capabilities of the wave analyser chiefly depend. For testing high-fidelity equipment it is necessary to measure distortion components of the order of $0.1 \%$ (i.e., 60 dB down) relative to a fundamental output separated in frequency by perhaps $25 \mathrm{c} / \mathrm{s}$, which relative to $50 \mathrm{kc} / \mathrm{s}$ is $0.05 \%$ off-tune. At the same time, in order not to render the analyser too tricky to use, or make unreasonable demands on signal-frequency stability, the filter characteristic ought to be flat-topped. Such onerous requirements, calling for carefully applied crystal resonator technique, have no doubt deterred many experimenters from running up an analyser for themselves.

Continuing the radio receiver analogy, we might remember that there is such a thing as a synchrodyne, ${ }^{2.3}$ which can be defined as a superhet in which the i.f. is zero. An advantage therein is that instead of the usual highly selective band-pass i.f. system a simple low-pass filter will do. This seems just what is wanted for a wave analyser. Providing a filter to cut out everything above one or two cycles per second presents no difficulty at all. In fact, on top of a
great saving in expense, it is easy to obtain substantially higher effective selectivity than in a conventional wave analyser, yet at the same time adjustment is less critical.

One inherent disadvantage of the principle ought perhaps to be declared at the outset. In a frequency changer, harmonics of the oscillator frequency are likely to be present. These, in a wave analyser with an i.f. of the order of $50 \mathrm{kc} / \mathrm{s}$, are too far up in the r.f. region to give i.f. beats with any a.f. (the " bias" frequency in a tape recorder must be watched, however). But in the zero-i.f. type of frequency changer, in which the desired response is obtained by setting the beat-oscillator frequency very nearly equal to that of the signal component to be measured, smaller responses can also occur at multiples (especially odd multiples) of the frequency read. One must therefore take some care to choose signal frequencies that cannot yield misleading responses.

What might be considered to be another disadvantage is that if one desires to cover the full a.f. band the beat oscillator has to be variable over a large tuning ratio, say $1000: 1$, as compared with less than 2:1 in the conventional system. On the other hand, however, the fact that the beat oscillator frequency at zero-beat is exactly the same as that of the component being measured makes for much easier adjustment and more accurate and stable frequency calibration, which are decided advantages.

Fig. 1 is a block diagram of a zero-i.f. wave analyser. When measuring very weak distortion products, some amplification ahead of the frequency changer could be helpful, but it is even more important not to add to the distortion before it is measured. Some sort of buffer stage is essential, in order to make the input impedance large enough not to impose appreciable loading on the signal being analysed and at the same time to present to the modulator an output impedance low enough not to introduce appreciable distortion. These requirements are met by a low-resistance cathode follower.

## Type of Oscillator

The modulator is a bridge comprising four rectifiers, shunted diagonally across the signal path. The design of the beat oscillator, which is connected across the other diagonal, depends on requirements. For experimental purposes an a.f. generator previously described ${ }^{4}$ was more than adequate, and enabled results using square waves to be compared with those using sine waves. For routine tests along the lines suggested, however, a simple fixed-frequency oscillator could be used. A 3 -valve oscillator like either of the two shown in Fig. 6 is very suitable for either fixed or variable frequency.

Because the mechanical movement of the indicator


Fig. 2. Circuit diagram of a particular development of Fig. I.
is itself a low-pass system, a very simple and cheap resistance-capacitance filter is all that is needed between modulator and amplifier. It was this amplifier that presented the greatest difficulty. At first is was supposed that two resistance-coupled stages with the time constants adapted to pass a band of the order of 0.2 to $2 \mathrm{c} / \mathrm{s}$, or a balanced amplifier with heavy negative feedback except over this band, would be a suitable basis for design. By cutting out zero frequency it was hoped to obtain a stable meter zero, but it was found that even in a push-pull system it was impracticable to balance the long time constants sufficiently to avoid perpetual slow drifts, and in spite of the zero-instability problem a d.c. amplifier was actually more workable. Stabilized power supplies are of course necessary in any case.

The design can now be discussed in detail with reference to the circuit diagram, Fig. 2, which is subject to modification to suit individual needs. In order to provide a low output resistance (about $130 \Omega$ ) at a moderate anode current ( 7 mA ) the cathode follower comprises both halves of a high-slope double triode. The distortion caused by it is too small to estimate precisely; but the total residual intermodulation, which with two input signals of different frequency, each $2 \frac{1}{2} \mathrm{~V}$ peak, is of the order of $0.05 \%$ is perceptibly increased if only one half is used. Without $\mathbf{R}_{8}$ and $R_{9}$ there is a tendency to parasitic oscillation. The time constant $C_{2} R_{5}$ has a valuable stabilizing effect in preventing short-term fluctuations in h.t. voltage from reaching the grid. $\mathrm{C}_{2}$ has in any case to be fairly large for its impedance at the lowest input frequency (reckoned as $20 \mathrm{c} / \mathrm{s}$ ) to be negligible in relation to $\mathrm{R}_{4}$, the lowest step of the input potential divider, which functions as a scale multiplier. Obviously the switch controlling it, $\mathrm{S}_{1}$, should be of the make-before-break type.

The modulator (known as a Cowan type ${ }^{2,3}$ ) is made up of four Standard Telephones M3 miniature selenium rectifiers. Under the influence of the beat oscillator their resistance becomes alternately much higher and much lower than $\mathrm{R}_{11}$, so that the signal from the cathode follower is alternately passed to the filter and suppressed. When the signal contains a frequency exactly equal to that of the oscillator, and in phase, the component of the signal at that frequency

Fig. 3. Signal waveforms across the modulator, (a) according to simplified theory, and (b) as modified in practice by $C_{3}$ (Fig. 2). The dotted portions of the original waveform are suppressed. In (b) the horizontal portions indicate the displaced zero level.
is in effect rectified. If the phase is reversed, the direction of rectification is reversed. So if the signal frequency is altered by, say, $1 \mathrm{c} / \mathrm{s}$, a $1-\mathrm{c} / \mathrm{s}$ signal is created by the slowly shifting phase relationship. This signal is accepted by the filter, the original signal being rejected.

## Modulation Wave Shape

If the action were quite as simple as just described, the waveform across the modulator, with the beat oscillator in phase with a sinusoidal incoming signal, would be as in Fig. 3(a), where the dotted line traces the suppressed half-cycles. This condition corresponds to the peak value of the beat-frequency signal, which is clearly equal to the mean value over a whole cycle of one half cycle of the incoming signal, or $1 / \pi$ times its peak value. The theoretical efficiency, neglecting all losses, such as those due to the finite resistance of the rectifiers, would therefore be barely $32 \%$. In reality, however, $\mathrm{C}_{3}$ modifies the action in a manner analogous to that of the reservoir capacitor in an ordinary rectifier circuit, displacing the waveform upwards and increasing its mean value and therefore the efficiency. To obtain the greatest benefit from this, the beat oscillator should have a pulse waveform, so that (just as in the rectifier circuit) the charging of $\mathrm{C}_{3}$ is concentrated at the peak
of the input signal. After investigation it was decided that the increase in efficiency was not worth the trouble of providing the special beat-oscillator waveform and ensuring constancy of its positive/negative time ratio at all frequencies. The undesired responses are also more liable to be troublesome, because if the waveform ratio is $m: 1$ the only harmonics suppressed are integral multiplies of $m+1$. Using a 1:1 waveform thus suppresses all even harmonics. Both for this reason, and in order to obtain constant efficiency, it is advantageous to preserve an accurate $1: 1$ ratio. This is easier to do with a sinusoidal waveform than with a square; so although the square gives the quickest switch-over for a given peak value, and consequently slightly lower residual intermodulation than an equal sinusoidal voltage, the latter was chosen on balance. The resulting signal waveform is as Fig. 3(b), and the measured efficiency (inclusive of the cathode follower) $52 \%$. For comparison, the efficiency with a $4: 1$ oscillator waveform was $75 \%$.

The time constant $\mathrm{C}_{3} \mathrm{R}_{11}$ is important, because it must be long compared with the period of the lowest signal frequency and short compared with that of the highest beat frequency. Since the optimum $\mathrm{R}_{11}$ depends on the backward and forward resistances of the rectifiers, a suitable time constant is obtained by choice of $\mathrm{C}_{3}$.

The higher the oscillator voltage the greater the signal voltage that can be handled linearly, but of course it must be within the maximum rating for the rectifiers, which is 56 V peak inverse per rectifier, or 112 for the bridge, less an allowance for inequality of backward resistance. The signal voltage being read is limited by the amplifier to about 10 V peak, and for that a sinusoidal oscillation of 20 V r.m.s. is sufficient. To allow a margin of amplifier linearity, the maximum reading has been reckoned as that which is given by 6 V r.m.s. at the input with $\mathrm{S}_{1}$ at $\times 1$. Provided that $S_{1}$ is not moved to a more sensitive setting than that at which the strongest signal component is within the $6-V$ limit, the peak value of the whole signal at the modulator is not likely to exceed about 20 , and using 28 V r.m.s. ( 40 V peak) for modulation it has been checked that this amount of signal voltage does not appreciably affect the accuracy of reading (e.g.) a 4 mV component of it. It may, however, make it rather less easy to read, and 10 V peak is a more conservative limit. The mean rectified current in the transformer secondary at 28 V r.m.s. is about 0.5 mA , which is well below the rated maximum of 1 mA per rectifier. Linearity of the modulator is excellent.

## Meter Zero Stabilization

In theory, this type of modulator automatically balances out the rectified and oscillator-frequency voltages from the signal path, but owing to inevitable inequalities in rectifier characteristics there is in practice a residue of both. A reasonable amount of this has not been found to cause appreciable error, but even a very small residual fraction of the total rectified (z.f.) voltage is enough to displace the pointer considerably. That in itself could be taken up on the amplifier balancing adjustment $\left(\mathrm{R}_{24}\right)$, but unfortunately the z.f. residue tends to vary with oscillator frequency, so that every change of frequency necessitates readjustment of meter zero. This nuisance $c i n$ be more or less eliminated by (1) winding the
transformer in the manner recommended for impedance bridges-with the secondary in two identical halves, balanced with regard to earth (represented mainly by the inter-winding screen);* (2) using the balance control $\mathrm{R}_{14}$; and (3) connecting a capacitance $\mathrm{C}_{4}$ across one of the rectifiers. This capacitance, of the order of 100 pF , has most effect at high frequencies. The procedure is to adjust $\mathrm{R}_{14}$ so that switching on the beat oscillator at some medium or commonlyused frequency (say about $400 \mathrm{c} / \mathrm{s}$ ) causes no displacement of the meter on the most sensitive range; then to vary the oscillator frequency and try various values and positions of $C_{4}$ to minimize shift when the frequency is raised. One of the advantages of tests at a single frequency ${ }^{1}$ is that no special transformer or other precautions are needed. Also the advantages of a square modulating waveform can be obtained without most of the disadvantages.

The component values of the filter are not critical, but obviously $\mathrm{R}_{15}$ must be large compared with $\mathrm{R}_{11}$ if efficiency is not to be lost. As an example of the selectivity obtained by the simple means of this instrument, reading an input of 2.5 mV at any frequency is quite unaffected (except for a very slight vibration of the pointer) by the presence of $4,200 \mathrm{mV}$ (i.e., 65 dB stronger) only $30 \mathrm{c} / \mathrm{s}$ away.

## Negative Feedback

At first glance the amplifier circuit may look rather like that of the valve voltmeter described in the August 1954 issue, but whereas in that circuit the whole voltage gain was sacrificed in the interests of stability by connecting the output terminals straight back to the control grids of $V_{2}$ and $V_{3}$, here a large voltage gain is needed in order to be able to measure small distortions, and the only negative feedback is that which results from feeding the screen grids of $V_{2}$ and $V_{3}$ from the output terminals. This policy not only saves a special potential divider for the purpose, but it provides some degree of zero stabilization, and lowers the output resistance of the amplifier, giving a more stable calibration. The resistance of the $250-0-250 \mu \mathrm{~A}$ meter is $500 \Omega$, and the output resistance of the amplifier about $160 \Omega ; \mathrm{R}_{29}$ is used to bring the whole up to a level of $700 \Omega$, so that $\mathrm{R}_{30^{-}}$ $\mathrm{R}_{32}$ with the values stated give decade ranges by means of $S_{2}$.
$\mathrm{R}_{28}$ is used to set the valves to suitable working points, indicated by the total amplifier h.t. current being about 7 mA . $\mathrm{R}_{24}$ is used as a coarse balancing or zero-setting control and $\mathrm{R}_{25}$ the fine control. The additional balancing facility afforded by $R_{22}$ is not absolutely essential, but is quite helpful in arranging the best working condition.

Without $\mathbf{R}_{19}$ and $\mathbf{R}_{20}$ the system was found to give full-scale swing on the $\times 1$ ranges of $S_{1}$ and $S_{2}$ for 9 mV r.m.s. input at the selected frequency. By means of $\mathrm{R}_{20}$ this is pre-set to a convenient 10 mV , the scale having been provided with a $10-0-10$ marking. On the $\times 1,000$ setting of $S_{2}$ the meter is therefore direct reading in volts, but (as already explained) should not be used above 6 on this range.

The overall voltage gain of the amplifier is thus about $\times 24$. Although not large by higher-frequency standards it can cause quite a lot of trouble unless care is taken with regard to zero stability. The balanced circuit of course goes a long way towards

* A suitable transformer can be obtained from the Majestic Winding Co., 180 Windham Road, Bournemouth.
achieving this and those inexperienced with d.c. amplifiers may perhaps wonder why it does not go all the way, since it might appear that any change in supply voltages would affect both halves of the system equally and therefore would not affect the meter. But a little calculation shows that a perceptible displacement of the pointer-say $10 \mu \mathrm{~A}$-results from a difference between the anode currents of $V_{2}$ and $V_{3}$ of only about $0.05 \mu \mathrm{~A}$. Now although any initial difference can of course be corrected by $R_{24}$ or $R_{22}$, it would be too much to expect the anode currents of even a well-matched pair of valves to vary equally within $0.05 \mu \mathrm{~A}$ over a range of anode, screen and heater voltages.

Any reasonably effective stabilizer for the source of the 300 V h.t. should be able to eliminate zero-shift due to variations in anode and screen potentials of valves $\mathrm{V}_{1}-\mathrm{V}_{4}$, but the provision of $\mathrm{R}_{22}$ in addition to $R_{24}$ makes assurance doubly sure; by successive!y shifting the setting of one and recentring the meter pointer with the other, an adjustment can be found at which even several volts change in h.t. has little effect on the meter, and protection is thus obtained against transient fluctuations of voltage of an inferior stabilizer. Protection against such transients (but of course not against long-term drifts) via the grids of $V_{1}$ is given by $C_{2} R_{5}$. Trouble due to small fluctuations of h.t. is thus confined almost exclusively to the cathode current of $V_{1}$, and it is for this that stabilization of the supply is chiefly needed. Even here an exceptionally high standard is not essential, thanks to the stabilizing effect of the negative feedback.

Incidentally, in case it occurs to anyone to extend the balanced-circuit principle all the way, as in Fig. 4, it should be mentioned that this causes the full signal voltage from $V_{1}$ to be applied between the grids of $V_{2}$ and $V_{3}$ and earth, thereby prematurely overloading


Fig. 4. Alternative, but less satisfactory, arrangement of the circuit up to the d.c. amplifier.
the amplifier on its most sensitive range. Moreover differences in time constants cause h.t. fluctuations to be more, rather than less, troublesome.

Although changes in oscillator amplitude affect the meter by upsetting the adjustment of $\mathrm{R}_{11}$, it has not been found necessary to stabilize the oscillator power supplies, sufficient control being provided by the thermistor. ${ }^{4}$

Except for very low frequency fluctuations that may come in with the signal, the remaining cause of zero instability is the heater voltage. Its effect on $V_{4}$ is slight. On $\mathrm{V}_{1}$ it is more troublesome; a $5 \%$ change in $V_{h}$ was found to cause $100 \mu \mathrm{~A}$ zero shift on the most sensitive range. But the effect is most serious as regards $V_{2}$ and $V_{3}$. The simple device of differentially adjusting the heater voltages of the two valves, for which very good results have been claimed with triodes, ${ }^{5}$ was found to be quite ineffective, at least as regards the few samples of EF86 available. If one is not so fortunate as the writer to pick two valves that are well matched, not only as regards the ultimate


Fig. 5. Cherry and Wild stabilizer for the heoter supply to $V_{2}$ and $V_{3}$, if found necessary. The VRI50 tubes in parallel are connected anode to cothode.
inequality in anode-current change caused by a heatervoltage change, but also in the rate at which the anodecurrent change occurs, it is advisable to stabilize the heater voltage of at least these two valves. If it can be stabilized for all the heaters, so much the better; but stabilization of $V_{2}$ and $V_{3}$ only is suggested as a secondbest, because the comparatively simple Cherry and Wild system ${ }^{6}$ provides sufficient output for these two heaters. Fig. 5 shows the circuit, which was found to be amply effective for badly matched valves, giving a stabilization ratio of the order of $30: 1$.

## Type of Meter.

Heater-voltage stabilization is rendered ineffective if there is instability of heater-circuit resistance, so good valve-holder contacts are essential; it would be better, if suitable valves with wire leads were obtainable, to solder them in without holders. Of course the whole of the circuit, especially from signal input to the grid of $V_{2}$, must be free from uncertain contacts, leakages and stray pick-up; for example, all capacitors should have high and constant insulation resistance, and their cases should be earthed.

Lastly, there is the indicator itself. A 2-in dia.
$0-500 \mu \mathrm{~A}$ meter, adjusted to a centre zero, was found to have a suitable combination of electrical and mechanical characteristics. Owing to the automatic limitation of current by the circuit, a more sensitive instrument could be used without risk of accidental damage, but there would be no point in doing so, because the minimum effective reading is already limited more by instability of zero and residual distortion than by insensitivity of the meter. Damping of the movement must be enough to suppress resonance without causing excessive sluggishness. The higher the frequency at which the pointer can oscillate without serious error ( $10 \%$ or even $20 \%$ is a not unreasonable tolerance in distortion measurements), up to the rate at which reading its amplitude of swing becomes difficult for the eye, the less critical is the adjustment of the signal frequencies and the demands on oscillator frequency stability. In practice a beat period of several seconds per cycle is about right.

## Intermodulation Twin Oscillator

Coming now to the signal source, the chief requirement is stability and ease of adjustment of frequency. Dependable frequency calibration is a great help, particularly in identifying and avoiding spurious responses. For measuring harmonics, very pure waveform is obviously needed; this is not so necessary with intermodulation, the errors caused by harmonic content then being only of a second order of magnitude. The three-valve circuit forming the nucleus of the a.f. source already mentioned has proved to be simple and reliable, with quite phenomenal frequency and amplitude stability. A double oscillator for intermodulation measurements, hastily put together without any special care, using ordinary components, has often been set to give a slow beat, at, say, $0.3 \mathrm{c} / \mathrm{s}$, and has continued to maintain this rate, without perceptible change, for hours on end. As regards purity of waveform, the second harmonic is about $0.25 \%$, third $0.3 \%$, and all others negligible. This performance was obtained notwithstanding that six old EF50 valves were substituted for the SP61 used in the original signal generator, whose harmonics are about half as great. The suitability of EF 80 valves was checked by plugging them in, via adapters, in place of the EF50; no readjustment was necessary; the total h.t. consumption for the two oscillators was increased from 40 mA to 46 mA , and the harmonic content nearly halved. In every way, therefore, this circuit is eminently suitable as a signal source for intermodulation measurements in conjunction with the wave analyser. Fig. 6 is the circuit diagram.

For economy as regards the ganged inverse semilog rheostats, the frequency ratio on each range was

TABLE I.

| Range | Oscillator 1 |  | Oscillator 2 |  |
| :---: | :---: | :---: | :---: | :---: |
|  | Frequency | Capacitance | $\begin{gathered} \text { Fre- } \\ \text { quency } \end{gathered}$ | Capacitance |
| 1 | 40-120 c/s | $0.1 \mu \mathrm{~F}$ | 350- | $0.0115 \mu \mathrm{~F}$ |
| 2 | 350- | $0.0115 \mu \mathrm{~F}$ | 1,100 c/s | $0.004 \mu \mathrm{~F}$ |
|  | 1,100 c/s |  | 3,200 c/s |  |
| 3 | $3-10 \mathrm{kc} / \mathrm{s}$ | 1,330pF | $3-10 \mathrm{kc} / \mathrm{s}$ | 1,330pF |
| 4 | $9-25 \mathrm{kc} / \mathrm{s}$ | 450 pF | $9-25 \mathrm{kc} / \mathrm{s}$ | 450 pF |

reduced from the $10: 1$ in the original model to about $3 \frac{1}{2}: 1$; this also slightly facilitated the all-important precision in frequency setting, but the fine controls $\left(R_{2}\right)$ are chiefly relied upon for close adjustment. For the purposes in view, complete coverage of the full a.f. band by both oscillators was unnecessary, and the ranges are as given in Table I.

If lower frequencies were included it would be necessary to increase the values of the coupling capacitors accordingly.

Apart from these modifications of the original design, concerned with frequency coverage, the only new consideration was the necessity for combining the two signals in the output, without giving rise to intermodulation. With the direct mixing system shown in Fig. 6, there is just perceptible intermodulation, and if one wants to avoid this one can substitute a bridge or hybrid-coil system, but the improvement was found to be hardly worth while. It can be shown that, given equal outputs from the two oscillators, the ratio of their voltages at the common output terminals is equal to the ratio in which $R_{19}-R_{21}$ is divided by $S_{2}$ (where these resistances include the internal resistances of the oscillators, which, being cathode followers in this case, are negligible). If therefore tests are usually made with either a $1: 1$ or 4:1 ratio, it is convenient to provide tappings as shown. This does not prevent any other desired ratio from being set up by the separate controls ( $\mathrm{R}_{14}$ ); the level of the combined signal can then be varied without change of ratio by $\mathrm{R}_{22}$.

## Avoiding Spurious Effects

The frequency-selecting section of each oscillator should be enclosed in a screen; it is particularly necessary to remove stray capacitance between the two sections, as failure to do so may cause a small signal from one oscillator to appear at the output even when its own $R_{14}$ is at zero. No difficulty was found, however, in running the twin oscillator unit from the same stabilized power supply as the wave analyser.

Residual intermodulation of modulator, input cathode follower, and signal-oscillator output circuit was checked in turn by first feeding the modulator with a single $420 \mathrm{c} / \mathrm{s}$ signal from oscillator 2 via analyser cathode follower in series with the secondary of a step-down mains transformer giving a $50 \mathrm{c} / \mathrm{s}$ signal, each signal being adjusted to give a reading of say 2 V on the analyser meter. Intermodulation signals were looked for at $420 \pm 50$ and $420 \pm 100 \mathrm{c} / \mathrm{s}$, and were barely perceptible on the most sensitive range. The $50 \mathrm{c} / \mathrm{s}$ source was then transferred to the lead between signal oscillator and analyser cathode follower. Lastly, the test was repeated with the $50 \mathrm{c} / \mathrm{s}^{\circ}$ mains source replaced by oscillator 1 . Overall intermodulation using a sinusoidal beat-oscillator waveform, was $0.1 \%$ second order and $0.08 \%$ third order. With a $1: 1$ square waveform the figures were $0.04 \%$ and $0.02 \%$ respectively.
Of spurious responses, the most important are those resulting from signals at multiples of the frequency being read, which is the frequency of the beat oscillator, denoted by $f_{b}$. Using a $1: 1$ waveform ratio, the responses due to signals at even multiples of $f_{b}$ are theoretically absent, and those at odd multiples (frequency $=n f_{b}$ ) are one $n$th of the amplitude resulting from an equal amplitude signal at $f_{b}$. In practice this is approximately true of the odd multiples, but although the even-multiple responses are relatively


Fig. 6. Circuit diagram of a twin signal source for making intermodulation measurements in conjunction with the wave analyser. The thermistors $\left(R_{f}\right)$ are Standard Telephones type A5513/100, and the volves all SP61, EF50, EF80 or equivalent. $S_{1}$ and $S_{3}$ are 2-pole 4 -way frequency-range switches; the capacitonces selected by pole b are the same as by pole a, and typical values are given in Table I. Components in Oscillator 2 having the same values as the corresponding components in Oscillator 1 bear the same numbers.
small ( $<5 \%$ of the true response) their existence ought not to be overlooked.

When analysing signals it is inevitable sometimes that the component being read will be accompanied by another component at a multiple of its frequency, in which case an error must result. Fortunately it is seldom large enough to be appreciable. One reason for this is that, taking the odd and even harmonic series separately, their amplitudes almost invariably decrease more or less rapidly with frequency; another is the decreasing response of the analyser, which almost wipes out the even series and greatly discriminates against the odd.

In the very unlikely event of a sixth harmonic not being small compared with a second harmonic, it might (being three times the frequency) appreciably affect the reading of the second. But the only troublesome situation that is at all likely in practice is the measuring of a fundamental accompanied by a very large percentage of third harmonic. This situation is revealed by the non-sinusoidal swinging of the pointer, for the third harmonic in the signal is represented (though at only one third the amplitude) by a third harmonic of the slow beat frequency, and the effect of this harmonic can be still further reduced by bringing the fundamental beat frequency to the point beyond which response falls off rapidly.

In choosing $f_{1}$ and $f_{2}$, the lower and higher signal oscillator frequencies for intermodulation tests, one would of course avoid making any of the components to be read- $f_{2} \pm f_{1}, f_{2} \pm 2 f_{1}$, etc.- equal to a multiple of $f_{1}$ or $f_{2}$. Fractional relationships between $f_{1}$ and $f_{2}$ would also be avoided; e.g., if $f_{1} / f_{2}$ were $2 / 5, f_{2}+2 f$ would clash with $f_{2}-f_{1}$, being its three-fold multiple. Doubt about whether or not a response is due to intermodulation can be dispelled by checking that the beat frequency responds to an adjustment of both $f_{1}$ and $f_{2}$.

Pick-up from the mains is another possible cause of undesired responses, and $50 \mathrm{c} / \mathrm{s}$ and its multiples (especially odd) are to be avoided as signal frequencies.

It has been interesting, in using this wave analyser, to confirm the experimental measurements by Warren and Hewlett ${ }^{\text {? }}$ of the ratio of intermodulation to
harmonics with amplifiers having various kinds of distortion and level frequency characteristics. Since the wave analyser reads only the sum or difference frequency and not both at once, the ratios are half those obtained by Warren and Hewlett with the same relative signal amplitudes (which were in the ratio 5 for harmonic to 4 and 1 for intermodulation measurement). With a single triode, giving almost pure second-order distortion, the amplitude of the component at $f_{2}+f_{1}$ or $f_{2}-f_{1}$ is 1.6 times that at $2 f_{1}$; and with a push-pull amplifier, giving mainly third-order distortion, the amplitude ratio of $f_{2}+2 f_{1}$ or $f_{2}-2 f_{1}$ to $3 f_{1}$ is 1.92 . When negative feedback is used, distortion is of course greatly reduced; but if the signal amplitude is then raised to reintroduce distortion, its onset is more rapid, and instead of being concentrated mainly in second or third or both, it continues far up the series with comparatively slow convergence, and the aural unpleasantness is worse.

While admittedly a keen ear is the only instrument that is valid in the final assessment of sound reproduction, instrument readings are of very great value if properly taken and interpreted; and if analysed measurements rather than lumped were more generally made the much-needed establishment of accepted correlation between instrument readings and listening tests would assuredly be speeded up.

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## TRANSISTOR Symbils

SUGGESTIONS for the symbol to be used for the transistor have appeared from time to time in Wireless World, and especially in the April and May issues (pp. 151 and 201). Another symbol of an interesting kind is used in L'Onde Electrique (March-April, 1955, pp. 243-263). The symbols for $n-p-n$ and $p-n-p$ junction transistors are shown in the figure, at (a) and (b) respectively. Plus signs are used to designate p-type material and minus signs for n-type.

Since the emitter, as their source, contains more charge carriers than the collector (by those leaving through the base), the emitter and collector are distinguished by two signs and one sign respectively.

The full sign complement of (a) and (b) is actually unnecessary, for the base material is always of opposite kind to the emitter and collector. One can, therefore, omit the signs on the base and reduce the symbols to the form (c) and (d). This is quite commonly done in the article in L'Onde Electrique. A further modification consists in replacing the signs by dots as in (e). This is done when one wants to designate a transistor without being specific about whether it is a $p-n-p$ or an $n-p-n$ type.

There is still some redundancy, however, and the symbols could be reduced to ( f ), ( g ) and ( h ) without

losing anything. The convention here is to mark only the emitter by a sign indicating the nature of the charge carriers, so (f) is for an $n-p-n$ transistor, ( $g$ ) is for a $p-n-p$ and ( $h$ ) is a general symbol for either.


PLANS have been made to build three new chains of Decca Navigator stations in Europe-in Sweden, southern France and northern Scot-land-making a total of nine. When the new stations are completed some $2,000,000$ square miles of Europe will be covered. As will be seen on this map, coverage will extend from Cape Finisterre to the Gulf of Bothnia and from Corsica to beyond the Faroes.

It is also announced that plans are being made for the first two chains of permanent stations to be erected outside Europe. They will cover the Bombay and Calcutta areas of India. Temporary lowpower chains have been set up overseas for survey purposes; one of the latest being for the Japanese Hydrographic Department.

Since the first Decca chain in south-east England was opened in July, 1946, over 2,600 naval and merchant ships have been fitted with the Decca Navigator. An increasing number of civil aircraft are also using it, and a new receiver, the Mark 10, designed especially for aircraft, is being produced. It will cover seventeen different chain frequencies and provide automatic lane and zone identification.

## LETTERS TO THE EDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

## Spurious Radiations from Wrotham

IN his article in your July issue J. R. Brinkley drew general conclusions on the co-siting of f.m. transmitters from observations made on the performance of the transmitters at Wrotham as they are at present installed. The arguments against co-siting of television transmitters are of course irrelevant when applied to f.m. transmitters, and "co-siting" is perhaps a misleading term to apply to the practice of feeding one aerial system with the output from three transmitters.

Mr. Brinkley is correct in pointing out that adequate steps must be taken to reduce coupling between transmitters, and he is also right in his statement that at the present time the attenuation of the filters now installed at Wrotham is insufficient. But, because of the severe interference on the medium-wave band the B.B.C. decided, in agreement with the Post Office and the radio industry, to bring Wrotham into regular service before the second half of the Home Service transmitter and the aerial combining units for the three transmitters were in their final form. In this way a v.h.f. service was made available at the earliest possible moment to listeners in London and South-East England. The temporary filters at present in use will be replaced by the final filters within the next few weeks. When this has been done the intermodulation products to which Mr. Brinkley refers, and of which we are naturally aware, will become negligible.
E. L. E. PAWLEY

Head of Engincering Services Group, B.B.C.

## F.M. Receiver Design

WE cannot agree with J. K. Carter (correspondence in July issue) who rebukes us for using the ratio detector in our f.m. tuner described in the April and May, 1955, issues. The decision to use this in preference to the Foster-Seeley discriminator was made after considerable thought and the saving of one valve is only one of the factors which influenced us. A more important consideration is that the low distortion of the Foster-Seeley circuit can be obtained only by critical adjustment of the coupling between primary and secondary windings of the discriminator transformer. To make this adjustment requires equipment unlikely to be possessed by the amateur constructor. The linearity of the ratio detector is less dependent on circuit adjustment; moreover there is less inter-station noise than with the Foster-Seeley type.

Mr . Carter accuses us of being illogical but he conveniently overlooks the other sources of distortion in the complete f.m. chain and in particular the chief offender; namely, the moving-coil loudspeaker. "Or," to use his own words. "is there some mystic reason why $n \%$ distortion in the loudspeaker doesn't matter but $n \%$ cent in the output stage does?"
S. W. AMOS, G. G. JOHNSTONE.

## Design for a Pre-amplifier

D. H. W. BUSBY, on p. 328 of his article in your July issue, says that a capacitance of 400 pF could be placed across the output with the gain control fully advanced, with negligible loss of output at $15 \mathrm{kc} / \mathrm{s}$.
This may well be so, but if the gain control is turned down to half-way, the reactance of the capacitor would appear across $50 \mathrm{k} \Omega$ and would be fed from a source impedance of at least $50 \mathrm{k} \Omega$. In this case the response would be down by at least an extra 3.5 dB at $15 \mathrm{kc} / \mathrm{s}$. In addition, the amount of high-frequency cut introduced would vary with gain control setting, being greatest with the slider electrically at centre, and would become progressively less on either side of this position.
London, S.E. 26 .
W. C. R. WITHERS.

The designer of the pre-amplifier zwrites.-Mr. Withers is quite correct. The amount of high frequency cut will
indeed vary with the gain control setting. If the loss due to the interconnecting cable is not to exceed 1 dB at $15 \mathrm{kc} / \mathrm{s}$ for any setting, the corresponding permissible value of capacity is approximately 150 pF . This will normally correspond to at least $7-8 \mathrm{ft}$ of cable, which, for most purposes, will be found adequate.

D. H. W. BUSBY.

## Damping Factor : A New Approach

AS the one originally responsible for introducing the term "damping factor "» the writer feels some responsibility for finding an alternative form now that we are so deeply in the morass. The term had many shortcomings but it could, at least, be used safely so long as it was always finite and positive. The commercial release of amplifiers with negative damping factors has been very confusing to engineers, to say nothing of the general public. For an increase of $22 \%$ in total circuit damping, the "damping factor" increases from 10 to infinity, then returns back from -infinity to -10. All these extraordinary changes in the damping factor would lead one to believe that something important was happening. In reality nothing has happened except a slight and steady increase in the total damping. The tricks played by the so-called damping factor are due merely to an unfortunate choice of definition. With this definition, instability occurs when the damping factor $\leq-1$.

The total circuit damping is a function of the total circuit resistance, that is, the algebraic sum of the voice coil resistance (always positive) and the amplifier output resistance (positive or negative). I therefore put forward the following as a much more satisfactory and logical substitute for damping factor:

$$
\text { Damping ratio }=\frac{\mathbf{R}_{\mathrm{L}}}{\mathbf{R}_{\mathrm{L}}+\mathbf{R}_{\mathrm{O}}}
$$

## Where

$\mathrm{R}_{\mathrm{L}}=$ load resistance
$\mathrm{R}_{0}=$ output resistance of amplifier and where both $\mathrm{R}_{\mathrm{L}}$ and $\mathrm{R}_{0}$ are referred to the same side of the transformer.

The following table is for $\mathrm{R}_{\mathrm{L}}=15$ ohms and is purely as an example:

| $\mathbf{R}_{0}$ <br> ohms | $\mathbf{R}_{\mathrm{L}}+\mathbf{R}_{0}$ <br> ohms | Damping <br> factor <br> $=\mathbf{R}_{\mathrm{L}} / \mathbf{R}_{0}$ | Damping <br> ratio <br> $=\mathbf{R}_{\mathrm{L}} /\left(\mathbf{R}_{\mathrm{I}}++\right.$ <br> $\left.\mathbf{R}_{0}\right)$ |
| :--- | :---: | :---: | :---: |
| -75 | +90 | +0.2 | 0.167 |
| $+\mathbf{3}$ | +18 | +5 | 0.83 |
| +1.5 | +16.5 | +10 | 0.91 |
| +0.15 | +15.15 | +100 | 0.97 |
| 0 | +15.0 | $\infty$ | 1.0 |
| -0.15 | +14.85 | -100 | 1.01 |
| -1.5 | +13.5 | -10 | 1.11 |
| -5.0 | +10.0 | -3 | 1.5 |
| -12.0 | +3.0 | -1.25 | 5.0 |
| -13.6 | +1.4 | -1.1 | 10.7 |
| -14.3 | +0.7 | -1.05 | 21.4 |
| -15.0 | 0 | -1.0 | $\infty$ |
| on verge of instability |  | $\infty$ |  |

It will be seen that the proposed damping ratio is positive and finite so long as instability does not occur. It is also proportional to the actual damping in the circuit. It appears to be the only available function with all the desired qualities.

Amalgamated Wireless Valve Company, Sydney, N.S.W., Australia.

[^1]
# Band II F.M. Tuner Unit 

Design Suitable for Use With a Wide Range of A.F. Amplifiers

A
LTHOUGH the tuner circuit described in the following pages was designed primarily for use with the Mullard 5 -valve 10 -watt amplifier circuit ${ }^{1}$, or with the 20 -watt circuit using EL34s ${ }^{2}$, it is suitable for use with a wide range of amplifiers. The frequency range covers the whole of Band II (87.5$100 \mathrm{Mc} / \mathrm{s}$ ), and while the circuit design chosen incorporates some of the more modern developments applicable to this type of reception, the construction is kept free from complication. The power supply would normally be taken from the main audio amplifier.
Circuit Description.-A complete circuit of the tuner unit is given in Fig. 1. The r.f. stage is a conventional pentode amplifier circuit, using a Muliard EF85. The associated aerial circuit is pre-tuned to the centre of the Band II range. In the anode circuit is an r.f. choke ( $\mathrm{L}_{3}$ ), with the parallel-fed r.f. tuned circuit ( $L_{4}, C_{9}, C_{10}, C_{11}$ ), of which $\mathrm{C}_{10}$ forms one section of the two-gang tuning capacitor. The r.f. voltage, taken from a tap on $\mathrm{L}_{4}$, is fed into the grid circuit of a Mullard EF80 operating as a selfoscillating additive mixer. The oscillator section of the mixer is basically of the tuned-anode type, with a tuned circuit consisting of $\mathrm{L}_{6}, \mathrm{C}_{12}, \mathrm{C}_{13}$ and $\mathrm{C}_{14}$ ( $\mathrm{C}_{13}$ forming the second section of the tuning gang.) The intermediate frequency developed by the mixer valve is fed to two conventional i.f. stages using Mullard EF41s, at an intermediate frequency of $10.7 \mathrm{Mc} / \mathrm{s}$. In the second i.f. stage the EF41 operates as a partial limiter valve at high signal levels, and the bias developed across its associated grid circuit capacitor $\mathrm{C}_{32}$ is fed back to the first i.f. valve and to the r.f. valve. The final i.f. stage drives a ratio detector circuit containing a Mullard double
diode, type EB91. Audio-frequency voltages developed in the detector circuit are taken through a 50 -microsecond de-emphasis network consisting of $\mathrm{R}_{16}$ and $\mathrm{C}_{45}$, and thence to the a.f. output socket for feeding to the audio amplifier.

Aerial Circuit and R.F. Stage.-The aerial circuit has been designed to be matched to a $75-\Omega$ balanced feeder line, thus permitting a simple connection from a conventional type of dipole aerial. In cases where the feeder line of this type is unscreened, $\mathrm{L}_{1}$ may be centre-tapped to earth so that any noise voltages picked up in the feeder itself are reduced. Dust core tuning is employed, and the resonant frequency of the grid tuned circuit is arranged to be $94 \mathrm{Mc} / \mathrm{s}$. The total tuning capacitance in the grid circuit is of the order of 18 pF , of which $\mathrm{C}_{2}$ forms 5 pF and the remainder is formed by the valve input capacitance, plus stray capacitance. The input damping of the valve amounts to $3,800 \Omega$, giving an effective secondary circuit impedance of the order of $1,600 \Omega$ (without the aerial circuit connected). When attached to an appropriate feeder cable, the aerial circuit bandwidth is $10.8 \mathrm{Mc} / \mathrm{s}$ for 3 dB down on $94 \mathrm{Mc} / \mathrm{s}$ and the measured aerial gain is 14 dB .
$\mathrm{C}_{10}$ which has a maximum capacitance of 27 pF tunes the r.f. circuit, and the series capacitor $\mathrm{C}_{9}$ and trimmer $\mathrm{C}_{11}$ are added to track the r.f. circuit correctly to the oscillator circuit. Thus the equivalent capacitance swing is limited to approximately 8 pF , and, in addition to the lumped capacitor constants, a further 12 pF and 4 pF are added to the circuits in the form of the r.f. valve output capacitance plus strays and the equivalent input capacitance of the mixer reflected into the tuned circuit, respectively. To assist further in obtaining correct tracking over

Fig. I. Complete circuit diagrom. Further details of component specification are given at the end of the article.


By

## L. IIAMPSON, B.Sc*

the whole of the band, $L_{4}$ is tapped by means of $\mathrm{C}_{15}$, so that the loaded Q factor of the r.f. circuit is comparatively high. The average value is 75 , and a mean bandwidth of 1.3 $\mathrm{Mc} / \mathrm{s}$ for 3 dB attencation is maintained over the whole tuning coverage, for any point in the band. Therefore, with an equivalent load impedance of the order of $4.5 \mathrm{k} \Omega$ at $94 \mathrm{Mc} / \mathrm{s}$ for the r.f. tuned circuit, a theoretical gain of 43 dB is obtained for the r.f. stage (including the aerial circuit). Under the circuit conditions shown the EF85 operates at a mutual conductance in the region of $9.5 \mathrm{~mA} / \mathrm{V}$. Measurement of the stage gain showed it to be only slightly less than calculated.

Mixer Circuit.-The self-oscillating type of mixer adopted in this circuit, has proved highly popular in countries where f.m. reception is well established. For successful operation the frequency difference between the developed intermediate frequency and the incoming signal frequency should be large. Fortunately this is generally so in f.m. receiver applications, where the intermediate frequency is usually about $10 \mathrm{Mc} / \mathrm{s}$. As its name implies it is essentially an oscillator with provision for feeding in an r.f. signal, so that additive mixing occurs on

[^2]
a common electrode (in this case the control grid of the EF80). The anode circuit contains an i.f. transformer which is tuned to the intermediate frequency developed.

The voltage gain of this single valve circuit is equivalent to that of a two-valve stage consisting of an oscillator and a separate mixer. At the same time the inherently low equivalent noise resistance obtainable with additive mixing is retained. In addition the use of a single valve makes the circuit economically more attractive.

In a mixer of this type, it is generally essential to have some form of "isolation" between the r.f. tuned circuit and the oscillator circuit of the mixer to prevent interaction and pulling of the oscillator section. This is achieved by operating the oscillator in a bridge circuit, the equivalent circuit of which is shown in Fig. 2. It will be seen then, that for the bridge circuit to be in balance, the relation,

$$
\mathrm{C}_{16}=\frac{\mathrm{C}_{\mathrm{in}} \times \mathrm{L}_{5 \mathrm{~A}}}{\mathrm{~L}_{5 \mathrm{BB}}}
$$

must hold. When this condition has been achieved there will be minimum interaction between the two relevant circuits. An obvious added advantage of this bridge connection is that when in balance, there will be minimum oscillator voltage at the r.f. input point. This is important in order


Fig. 2. Equivalent bridge circuit of oscillator section of mixer valve. $C_{i n}$ represents the valve input capacitance.
to keep the oscillator voltage at the aerial terminals as low as possible. In this circuit, it is possible to reduce the oscillator voltage to an average value of 100 mV at the r.f. input point.

The oscillator circuit is operated at a frequency higher than the signal frequency, i.e. at approximately 97 to $111 \mathrm{Mc} / \mathrm{s}$. As with the r.f. circuit, a series capacitor $\mathrm{C}_{12}$ and a parallel trimmer $\mathrm{C}_{14}$ are included for tracking purposes, giving an effective swing of the tuning capacitor $\mathrm{C}_{13}$ of approximately 7 pF . To keep the oscillator circuit as stable as possible the total capacitance associated with the tuned circuit has been made as large as possible, consistent with obtaining sufficient oscillator drive for the mixer valve. Also the cathode of the mixer valve is directly earthed in order to avoid capacitive hum modulation.

The blocking capacitor $\mathrm{C}_{18}$ also forms the tuning capacitance for $L_{7}$, since $L_{6}$ is effectively a short-circuit at the intermediate frequency. Similarly $\mathrm{L}_{7}$ forms an r.f. choke at the oscillator frequency. It will be seen then that the effective lumped tuning capacitance across $L_{7}$ is equivalent to:

$$
\frac{\mathrm{C}_{22} \times \mathrm{C}_{18}}{\mathrm{C}_{22}+\mathrm{C}_{18}}
$$

and the voltage tap down in the i.f. transformer is equal to
$\frac{\mathrm{C}_{22}}{C_{\text {out }}+\mathrm{C}_{18}+\mathrm{C}_{22}}$ where $C_{\text {out }}$ is the valve output capacitance.
However, as the value of $\mathrm{C}_{22}$ is so much higher than $\mathrm{C}_{18}$, the loss in gain of the mixer is negligible.

In order to ensure that the mixer valve operates on the optimum point of the conversion conductance curve, it is recommended that the oscillator grid


Fig. 3. Tuning characteristics for various input signal levels at the aerial terminals, with a frequency deviation of 22.5 $\mathrm{kc} / \mathrm{s}$.
current should not fall below a level of $35 \mu \mathrm{~A}$ in any part of the band. It will be found in operation that the oscillator grid current will not vary by more than approximately $\pm 10 \%$ of the mean value over the tuning range.

A conversion conductance of $2.5 \mathrm{~mA} / \mathrm{V}$ is obtainable with the EF80. Therefore, with the i.f. transformer used, the mixer gain from the tap point of $L_{5}$ to the signal grid of the first i.f. valve will be 32 dB . Due to the tapping of $\mathrm{L}_{4}$, the effective gain of the r.f. stage is proportionately reduced to 34 dB . At $94 \mathrm{Mc} / \mathrm{s}$ the overall " front end" gain is of the order of 66 dB , measured from the aerial terminal to the control grid of the first i.f. valve.
I.F. Stages.-The design of i.f. stages for f.m. receivers presents a considerable number of conflicting problems. Very briefly summarized they are:-
(i) There should be adequate transmission of the significent side currents.
(ii) A good measure of adjacent channel selectivity is required.
(iii) There should be a reasonably linear phase/ frequency characteristic in the transformers.
(iv) A certain allowance in the bandwidth should be made for small random drift in the oscillator.
(v) The uses of comparatively large values of overcoupling to give a wider bandwidth can produce a high degree of amplitude modulation on the carrier wave, and may give rise to ringing with impulsive interference.

It was decided that for the two i.f. transformers used in this tuner, a coupling factor K of design centre 1.2 , would be most suitable to meet a compromise for the above requirements, provided the average loaded $Q$ factor of the tuned circuits in the i.f. transformers is in the region of 60 to $70 . \quad\left(\mathrm{K}=k \sqrt{\mathrm{Q}_{\mathrm{p}} \mathrm{Q}_{\mathrm{s}}}\right.$, where $k$ is the coupling coefficient and $Q_{p}$ and $Q_{8}$ refer to primary and secondary windings.)

In addition to the above requirements it is essential that feedback through the anode-to-grid capacitance of the valve, should be kept small. This usually calls for a strict limit on the maximum usable transfer impedance obtained in an i.f. transformer when applied with any particular valve type. With the EF41, the recommended maximum design centre transfer impedance for the i.f. transformer is $21 \mathrm{k} \Omega$ at 10.7 $\mathrm{Mc} / \mathrm{s}$, taking into account the added effective anode-to-grid capacitance in the valveholder, and assuming identical impedances in grid and anode circuits.

Table 1 gives a summarized performance of the transformers used in this tuner. All the values quoted were measured in circuit.
The two EF41 valves operate with a mutual conductance of $2.3 \mathrm{~mA} / \mathrm{V}$, and from the relevant impedances given in Table 1, it can be calulated that the total i.f. gain, from the control grid of V3 to the anode of V4 will be 64 dB . Actual measurement showed a slightly lower value.

Measurement of the overall bandwidth of the first and second i.f. transformers gave approximately $210 \mathrm{kc} / \mathrm{s}$ for 3 dB , and $600 \mathrm{kc} / \mathrm{s}$ for 20 dB , attenuation. The response curve is flat for approximately $70 \mathrm{kc} / \mathrm{s}$, with a slight dip at the centre frequency.

The coils for the i.f. transformers are wound on common formers, details of which are given later in the Appendix. An inherent drawback in this method of construction is that movement of the dust cores can materially alter the coupling factor if the primary and secondary windings are, by necessity, brought too close together. This is to some extent eliminated in the design presented here, by separating about $25 \%$ of the total windings on the coils, so as to form small coupling coils at the earthy end of each winding. Thus the dust cores are kept at least 15 mm distant in the main body of the windings and little measurable
difference in the coupling factor is obtained when the transformers are tuned to $\pm 1 \mathrm{Mc} / \mathrm{s}$ of the correct working centre frequency. The unloaded $Q$ factors of the coils are about 90 , with the exception of $L_{12}$ which is approximately 85 .

As previously mentioned, on high signal levels V4 is driven into appreciable grid current, and the derived bias voltage developed across $\mathrm{R}_{11}$ and $\mathrm{R}_{12}$ is used as semi-automatic gain control. Positive peaks of amplitude modulation appearing on the carrier wave are therefore clipped in the grid circuit. With V4 primarily designed as an amplifier, a very high input voltage to the grid is required before saturation occurs. Therefore with only partial limiting occurring, the valve does not deal completely with positive-or negative-going amplitude variations of the carrier. For similar reasons partial limiting may give only restricted elimination of impulsive interference. The time constant of the grid circuit of V 4 has been set at 25 microseconds to deal with the upper frequency limits of amplitude variations in the carrier. The small changes in the input capacitance of V4 due to limiter action in the grid circuit and of V3 due to the restricted range of applied bias voltage were not found to introduce any serious deterioration in the required performance of the band-pass filters. Further, harmonics produced by limiter action do not usually cause trouble, as they will be tuned out in the anode circuit of V4.

Ratio Detector-A balanced ratio detector circuit incorporating a double diode (V5) is used in this tuner. To give optimum a.m. rejection under working conditions, $\mathrm{R}_{17}$ should be adjusted for each individual circuit and is specified here as a nominal value.

The value of a.m. rejection measured in the circuit was 46 dB at the centre frequency ( $10.7 \mathrm{Mc} / \mathrm{s}$ ) for 30 V r.m.s. of i.f. voltage at the anode of V 4 , falling to

View of underside of chassis with some of the more prominent components identified.


Wireless World, AUgust 1955

28 dB for $+75 \mathrm{kc} / \mathrm{s}$ detuning of the signal. Similarly for 20 and 10 V r.m.s. of i.f. signal, the values were 34 and 26 dB respectively, with corresponding values of 22 and 17 dB for $\pm 75 \mathrm{kc} / \mathrm{s}$ detuning. These figures include limiter action produced by V4. The peak separation of the " $S$ " shaped detector curve is $320 \mathrm{kc} / \mathrm{s}$.

Measurement of the tuning characteristic is advantageous in a prototype f.m. receiver, in order to examine the side responses which are inherent in most f.m. detector systems. These are shown in Fig. 3, where the tuning characteristic of the unit is plotted for various values of input signal.

These characteristics can be regarded as typical for a receiver equipped with a ratio detector circuit. The side responses are shown at about $\pm 200 \mathrm{kc} / \mathrm{s}$ from the centre frequency. To a certain extent these side responses are controlled by the selectivity characteristic of the preceding i.f. stages, and also by the action of a.g.c. In general, receivers designed with a rounded-top overall i.f. response curve, and a comparatively wide peak separation in the detector system help to reduce side responses and produce a peak in the audio output, when the signal is in tune. It may be noted in passing that, with the Foster-Seeley detector circuit and limiter valves, the side responses may be of a higher value than the main signal response. Although the side responses show a comparatively high value in the graph at the larger signal levels, they are in fact hardly noticeable when tuning through the signal.

The audio output from the detector is taken through the 50 -microsecond de-emphasis network $R_{16}$ and $C_{45}$, and coupling capacitor $C_{46}$ to the a.f. output socket. When the output from the tuner is fed to an audio amplifier it is recommended that the amplifier input impedance be not less than $500 \mathrm{k} \Omega$. For use with the pre-amplifier ${ }^{3}$ designed for use with the 20 -watt EL34 circuit ${ }^{2}$ and with other pre-amplifiers of similar input impedance and sensitivity it is recommended that a correction circuit (Fig. 4) be used to obtain the required input impedance and attenuation.

Tuning Indicator.-An optional tuning indicator using a Mullard EM80 is fitted. The bias voltage for this valve is derived from the ratio detector circuit.

Overall performance.- The total gain of the prototype tuner from the aerial terminals to the anode of V4 was approximately 130 dB for a small signal at 94 $\mathrm{Mc} / \mathrm{s}$. When coupled to a Mullard 5 -valve 10 -watt amplifier, the average sensitivity over the band for 50 mW output was $1.2 \mu \mathrm{~V}$ with a signal of $22.5 \mathrm{kc} / \mathrm{s}$ deviation. The average input signal over the band for 500 mV audio output is approximately 12 to $15 \mu \mathrm{~V}$.

Oscillator radiation. - The average oscillator voltage measured at the aerial terminals was $350 \mu \mathrm{~V}$ for the fundamental oscillator frequency and $75 \mu \mathrm{~V}$ for the 2 nd harmonic. The average radiated field strengths over the band are $40 \mu \mathrm{~V}$ per metre and $<15 \mu \mathrm{~V}$ per metre respectively at a distance of 10 metres from the measuring aerial.

Constructional details.-The accompanying photographs show the main layout of the tuner. A chassis of 16 s.w.g. aluminium, dimensions 10 in $3 \frac{1}{2}$ in, with 2 in depth, is used. With the exception of the valves, i.f. transformers, aerial and oscillator coils and scale assembly, all components are mounted underneath the chassis.
A balanced heater circuit is used. This enables the tuner to be connected to the centre-tapped heater supply of either of the amplifier circuits previously


Fig. 4. Simple attenuator recommented for use with amplifiers of high sensitivity and input impedance less than $500 \mathrm{k} \Omega$.
referred to ${ }^{1.2}$. Adequate r.f. decoupling of the heater supply lead is essential, in particular to prevent harmonics from the detector from reaching the earlier stages. The ninth harmonic of the i.f. can be particularly troublesome, as this falls in the centre of the tuning range. Strict attention should be paid to the general decoupling of the valve electrodes, and to the tuned circuits. The relevant decoupling capacitors, should be returned to, or very near to, the cathode-tochassis connection of the valve concerned, with the shortest possible leads. It is also essential that $\mathrm{C}_{16}$ should have a short connection to chassis.

The tuner requires an h.t. supply of 200 V at approximately 37 to 40 mA and a heater supply of 6.3 V at 1.6 A . Where the h.t. supply is obtained from an amplifier and exceeds 200 V the necessary dropping resistance should be included in the amplifier itself.

To minimize oscillator drift during the warming up period, $\mathrm{C}_{12}$ should be of the negative temperature coefficient type, a component of temperature coefficient 750 par:s per million being suitable. This will help to keep the long term oscillator drift to a minimum. In the prototype, $\mathrm{C}_{14}$ and $\mathrm{C}_{16}$ were formed of a 6.8 pF capacitor in parallel with a $1.25-10 \mathrm{pF}$ trimmer to make up the nominal total. The tap on the r.f. coil is arranged to be at 0.4 of the total number of turns, counting from the earthy end. Optional i.f. traps $\mathrm{L}_{14}, \mathrm{C}_{51}$ and $\mathrm{L}_{15}, \mathrm{C}_{52}$, have been incorporated in the aerial circuit for use where a high degree of i.f. rejection is considered essential. In most cases they may be found unnecessary.

Neatness in wiring is essential. In particular, in the i.f. stages, components should not be piled over the top of the valveholder as this may lead to an increase of effective anode-to-grid capacitance.

Alignment.-The correct alignment of an f.m. receiver calls for the use of some expensive equipment, but good results can be obtained by using only an a.m. signal generator, covering Band II (87.5-100 $\mathrm{Mc} / \mathrm{s}$ ) and the intermediate frequency of the tuner ( $10.7 \mathrm{Mc} / \mathrm{s}$ ). As the first two i.f. transformers are overcoupled, it is essential to damp the transformers whilst they are being tuned, otherwise unsymmetrical response curves may result. A resistor of about $5 \mathrm{k} \Omega$ is suitable for this purpose and it should be placed across the grid circuit tuned winding, when the anode circuit is being tuned, and vice versa. The resistor can be temporarily held on with a touch of solder.

Connect the signal generator output ( $10.7 \mathrm{Mc} / \mathrm{s}$ ) to the control grid of V 4 and tune $\mathrm{L}_{11}$ for maximum deflection either in the tuning indicator or on a high resistance voltmeter ( $20 \mathrm{k} \Omega / \mathrm{V}, 10 \mathrm{~V}$ scale) across $\mathrm{C}_{47}$. Transfer the generator in turn to the grid of V3, and the centre-tap point of $\mathrm{L}_{5}$. Tune $\mathrm{L}_{9}, \mathrm{~L}_{10}$ and $\mathrm{L}_{7}$ and $\mathrm{L}_{8}$ respectively for maximum deflection, using the damping resistor for the coils as before.

To eliminate considerable trial and error in the alignment of the r.f. and oscillator circuit, some approximate values of the correct trimmer settings
are given. $\mathrm{C}_{18}$ can be set initially at about 10 pF , and $\mathrm{C}_{14}$ to 12 pF , with the dust core of $\mathrm{L}_{6}$ tuning in the base end of the coil away from $\mathrm{L}_{3} . \mathrm{C}_{11}$ is set at approximately 5 pF . The bridge circuit may be balanced by connecting an r.f. valve voltmeter from the tap point of $L_{4}$ to earth, and adjusting $C_{16}$ for minimum oscillator voltage. While this operation is being done, the main tuning gang should be set with the vanes about half-way between minimum and maximum capacitance. If an r.f. valve voltmeter is not available, a rough and ready, but quite effective method of obtaining balance is to short-circuit the tap on $\mathrm{L}_{4}$ to earth and observe the change in grid current through $\mathrm{R}_{5}$ (on $50 \mu \mathrm{~A}$ scale). $\mathrm{C}_{16}$ is then adjusted until the change in grid current on shortcircuiting $\mathrm{L}_{4}$ to earth is a minimum.

With the signal generator connected at the aerial terminals and the tuning gang at maximum capacitance apply a signal of $87 \mathrm{Mc} / \mathrm{s}$, and tune $\mathrm{L}_{6}$ dust core so that a maximum deflection is indicated in the tuning indicator or voltmeter. Re-tune the signal generator to $100 \mathrm{Mc} / \mathrm{s}$ and adjust $\mathrm{C}_{14}$ for optimum output with the tuning gang at minimum capacitance. These settings may need to be checked a number of times to give the correct frequency range for the oscillator.

Re-set the signal generator to $91 \mathrm{Mc} / \mathrm{s}$ and adjust $L_{1}$ dust core for maximum output. Adjust $C_{11}$ for
maximum output at $98 \mathrm{Mc} / \mathrm{s}$ and finally set $\mathrm{L}_{2}$ dust core for maximum output at $93-94 \mathrm{Mc} / \mathrm{s}$. The dust core of the aerial coil should also be at the base end of the former. To align the i.f. traps apply a comparatively large input signal of $10.7 \mathrm{Mc} / \mathrm{s}$ to the aerial terminals, and tune $\mathrm{L}_{14}$ and $\mathrm{L}_{15}$ for minimum indicated output. With the signal generator connected again to V4 grid and with the signal of $10.7 \mathrm{Mc} / \mathrm{s}$ adjust $\mathrm{L}_{12}$ core, for zero d.c. voltage across $\mathrm{C}_{44}$. This ensures that the ratio detector circuit is reasonably well balanced.

## REFERENCES

1 " Mullard 5-valve 10-watt High Quality Amplifier Circuit," published by Mullard Ltd.
Also briefly described in:
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"A High-quality Ten-Watt Audio Amplifier," by D. H. W. Busby and W. A. Ferguson. Mullard Technical Communications. Vol. 1, No. 9. Nov., 1954.
2 "Design for a 20-watt High-Quality Amplifier, 2Constructional Details and Performance," by W. A. Ferguson. Wireless World, June 1955.
3 "Design for a Pre-amplifier, for use with a 20 -watt High-quality Amplifier." by D. H. W. Busby. Wireless World, July 1955.
(Appendix-Coil Winding Data-on next page)

## COMPONENTS LIST FÓR F.M. TUNER

| Capacitors |  |
| :---: | :---: |
| $\mathrm{C}_{1}$ | $120 \mathrm{pF} \pm 20 \%$ (C) |
| $\mathrm{C}_{2}$ | $5 \mathrm{pF} \pm 10 \%$ (C or SM) |
| $\mathrm{C}_{3}$ | $2,200 \mathrm{pF} \pm 20 \%$ (C) |
| $\mathrm{C}_{4}$ | $1,500 \mathrm{pF} \pm 20 \%$ (C) |
| $\left.\begin{array}{l} \mathrm{C}_{5}^{*} \\ \mathrm{C}_{6} \end{array}\right\}$ | $2,200 \mathrm{pF} \pm 20 \%$ (C) |
| $\mathrm{C}_{7}$ | 120 pF 上 $20 \%$ |
| $\mathrm{C}_{8}$ | $1,500 \mathrm{pF} \pm 20 \%$ |
| $\mathrm{C}_{9}$ | $33 \mathrm{pF} \pm 5 \%$ |
| $\left.\begin{array}{l} \mathrm{C}_{111} \\ \mathrm{C}_{13} \end{array}\right\}$ | $\begin{array}{ll}227 \mathrm{pF} & \text { Two-gang } \\ \text { (Jackson } & \text { U101 S-S }\end{array}$ variable |
| $\mathrm{C}_{11}$ | 3-15 pF (nominal) (composed of $1.25-10 \mathrm{pF}$ trimmer Wingrove \& Rogers, Type C32.01 $+6.8 \mathrm{pF}, \mathrm{SM}$ ) |
| $\mathrm{C}_{12}$ | $27 \mathrm{pF} \pm 5 \%$ (optional n.t.c. 750 parts per million) |
| $\mathrm{C}_{14}$ | 3-15 pF ( ( ${ }^{\text {aminal }}$ ) |

Resistors (all resistors $\frac{1}{2}$ watt Dubilier " BTS" type)
$\mathrm{C}_{15} \quad 120 \mathrm{pF} \pm 20 \%$ (C)
$\mathrm{C}_{16} \quad 3-15 \mathrm{pF}$ (nominal)
$\mathrm{C}_{17} \quad 22 \mathrm{pF} \pm 10 \%$ (C)
$\mathrm{C}_{18} \quad 22 \mathrm{pF} \pm 5 \%(\mathrm{C})$
$\left.\begin{array}{l}\mathrm{C}_{19} \\ \mathrm{C}_{20}\end{array}\right\} 2,200 \mathrm{pF} \pm 20 \%(\mathrm{C})$
$\mathrm{C}_{21} \quad 2,200 \mathrm{pF} \pm 20 \%$ (C)
$\mathrm{C}_{22} 0.01 \mu \mathrm{~F}$ Met. paper
$\mathrm{C}_{23} \quad 12 \mathrm{pF} \pm 5 \%$
$\mathrm{C}_{24} \quad 150 \mathrm{pF} \pm 20 \%$
$\mathrm{C}_{25}$
2,200 $\mathrm{pF} \pm 20 \%$ (C)
$\left.\mathrm{C}_{28}\right\} 0.01 \mu \mathrm{~F}$ Met. paper
$\left.\mathrm{C}_{31}\right\} 12 \mathrm{pF} \pm 5 \%$ (C or SM )
$\mathrm{C}_{32} \quad 47 \mathrm{pF} \pm 20 \%$
C - Ceramic. SM - Silvered mica

| $\mathrm{R}_{1}$ | $470 \mathrm{k} \Omega \pm 20 \%$ | $\mathrm{R}_{13}$ | $82 \mathrm{k} \Omega \pm 10 \%$ |  |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{2}$ | 1,200 ${ }^{\text {土 }}$ 20\% | $\mathrm{R}_{14}$ | $1,200 \Omega \pm 20$ |  |
| $\mathrm{R}_{3}$ | $68 \mathrm{k} \Omega \pm 10 \%$ | $\mathrm{R}_{15}$ | $47 \Omega \pm 10 \%$ |  |
| $\mathrm{R}_{4}$ | $15 \mathrm{k} \Omega \pm 10 \%$, | $\mathrm{R}_{16}$ | $100 \mathrm{k} \Omega \pm 10 \%$ |  |
| $\mathrm{R}_{5}$ | $100 \mathrm{k} \Omega \pm 10 \%$ $27 \mathrm{~S} \Omega \pm 10 \%$ | $\mathrm{R}_{17}$ | $1,200 \Omega \pm$ nom | al $\pm 5 \%$ |
| $\mathrm{R}^{6}$ | 3,300 ${ }^{2} \pm 20 \%$ | $\mathrm{R}_{18}$ | $2,700 \Omega \pm 5 \%$ |  |
| R8 | $1.0 \mathrm{M} \Omega \pm 20 \%$ | $\mathrm{R}_{19}$ | $\Omega+5 \%$ |  |
| R9 | $82 \mathrm{k} \Omega \pm 10 \%$ | $\mathrm{R}_{20}$ |  |  |
| $\mathrm{R}_{10}{ }_{10}$ | $1,200 \Omega \pm 20 \%$ |  | $470 \mathrm{k} \Omega \pm 20 \%$ | (For optional tuning in- |
| $\mathrm{R}_{11}$ | $470 \mathrm{k} \Omega \pm 20 \%$ $1.0 \mathrm{M} \Omega+20 \%$ | $\mathrm{R}_{22}$ $\mathbf{R}_{23}$ | $470 \mathrm{k} \Omega=20 \%$ | al tuning indicator) |

## Other Components

Miniature tag strips-British Moulded Plastics Type A 5556
Stand-off insulators-Wingrove and Rogers. Type TS101/1.
Scale and drive assembly-Jackson, Type SL15.

## Valves

V1 Mullard EF85
V2 Mullard EF80
V3, V4 Mullard EF41
V5 Mullard EB91
V6 Mullard EM80 (Optional tuning indicator)

## Coils

Aerial transformer $L_{1}, L_{2}$
R.F. choke $\mathrm{L}_{3}$
R.F. coil $L_{4}$

Oscillator coil $\mathrm{L}_{5}, \mathrm{~L}_{6}$
1 st i.f. coil $\mathrm{L}_{7}, \mathrm{~L}_{8}$
2nd i.f. coil $\mathrm{L}_{9}, \mathrm{~L}_{10}$
Ratio detector Transformer,
$\mathrm{L}_{11}, \mathrm{~L}_{12}, \mathrm{~L}_{13}$
I.F. $\operatorname{traps} \mathrm{L}_{14}, \mathrm{~L}_{15}$

Denco 510/AE
Denco 510/RFC
Denco 510/RF
Denco 510/OSC
Denco 510/IFT. 1
Denco 510/IFT. 2
Denco 510/RDT
Denco 510/IFF

APPENDIX—BAND II F.M. TUNER UNIT COIL WINDING DATA
I.F TRANSFORMERS
RATIO DETECTOR
TRANSFORMER

ALL COILS. EXCEPT $L_{12}$
WOUND WITH 35 s.w.

$L_{13}$
R.F. CHOKE

> 7TURNS OVERWOUNO ON LIIA SEPARATED FROM separated from Lila by 2 turns OF 0.001 in TRANSFORMER PAPER
$L_{\text {lia }}$
ITUARNS LIS
$L_{18}$
LIS
loturns


$$
\begin{gathered}
\mathrm{L}_{12} \\
14+14 \text { TURNS } \\
\text { BIFILAR WOUND } \\
34 \text { s.w.g. ENAM. }
\end{gathered}
$$

Formers: Aladdin PP No. 5937 (with screening can) or Neosid 5000B.
Dust Cores: Neosid Grade F 900 . Length 16 mm . Dia. 6 mm . Pitch 1 mm .

It is emphasized, that only the absolute minimum amount of sticky tape should be used to hold the i.f. transformer coils in position, otherwise the $Q$ factor will be affected considerably or the self-capacitance of the coils will be increased.


Formers: Aladdin PP No. 5938 (with screening cans) or Neosid 5000A.
Dust Cores: Neosid Grade I 901. Length 12.7 mm . Dia. 6 mm . Pitch 1 mm .

# New Navigational Aids 

## Radio and Radar Equipment at the Paris Air Show

THE twenty-first international aeronautical exhibition, held recently at Le Bourget aerodrome near Paris, demonstrated once again how very closely aviation progress and electronics are allied. Of the 180 exhibitors, representing ten different countries, some twenty were radio and radar manufacturers, and there was equipment from France, Great Britain, the U.S.A., Sweden, Italy and Germany on view.

In the field of radar, the French Compagnie Générale de Télégraphie sans Fil were giving interesting demonstrations of the remote presentation of radar pictures by a television system. The radar equipment was installed at Pontoise, north-west of Paris, and its images were transmitted by television link to the top of the Eiffel Tower, from which they were retransmitted to Le Bourget aerodrome. The system used depends largely on an analyser storage tube which the C.S.F. have developed and which is described as being a tube capable of storing signals in the form of a pattern of electrical charges deposited on a thin insulating target by a "writing" electron beam. The signals are "read" by a second beam, enabling them to be used to modulate a transmitter. The device is capable of storing the signals for periods ranging from microseconds to hours.

One half of the storage tube can be considered as the radar display system and has deflection coils synchronized with the aerial rotation. Instead of actually showing the image, however, it produces a charge pattern on the target plate in the middle of the tube. The electron beam in the other half of the tube, which has its own scanning system (usually a normal television scan), then discharges the target, the output of which is taken off at a collector electrode and used to modulate the television transmitter. The reading half of the tube is intensity modulated at $20 \mathrm{kc} / \mathrm{s}$ in order to avoid interference between the writing and reading scans. Thus on the writing side of the tube the scan can be radial, while on the reading side it can be orthogonal without inconvenience.

The television display equipment appeared to consist of domestic television receivers and they were relaying a radar picture of the Paris air traffic control zone. Until actual pictures are seen it is difficult to realize the full value of the storage tube for air operations, since the inherent tracking feature is so unusual to those who are used to seeing normal persistence radar screens. With the long "memory" the aircraft track is left on the screen from the moment it is first picked up until its goes out of range of the radar.

The chief advantages of the system are the inherent


Fig. 1. Rear warning radar aerial in the form of a horn in the tail of a French "Ouragan" fighter aircraft.
automatic tracking already mentioned and the great improvement in contrast obtained in the television reproduction of the radar image, which can be made far better than the original picture on the radar screen and clearly readable in daylight.

Another unusual radar system, known as the " 3 D " (for "three-dimensional"), was exhibited by the French firm Radio Industrie. The equipment, which operates in the $10-\mathrm{cm}$ band, provides range, bearing and height information from one aerial which moves only in the horizontal plane, the usual tilting movement of the reflector for height-finding being eliminated. This feature is obtained by the use of a system of scanning (known as a Robinson scan) in which the waveguide feed is moved mechanically so as to produce a change in the vertical direction of the beam from the aerial reflector. In practice the vertical beam can be varied over some 15 degrees and the scan rate is about 800 per minute. This form of height-finding is not new, of course, but in the Radio Industrie radar the disadvantages of previous equipments of this kind are claimed to have been overcome. The side-lobe suppression is stated to be greatly improved by the use of a specially shaped reflector and the beam width throughout the scan remains constant.

Coming now to airborne radar, an interesting feature of the cloud and collision warning radar shown by the Société Francaise Radioelectrique was the arrangement for preventing reduction in range by cloud "clutter" when the equipment is being used for collision warning. This takes the form of a series of probes which are introduced at an angle of $45^{\circ}$ into the waveguide feeding the reflector. These are stated to make the polarization circular, under which conditions echos from clouds and rain are greatly reduced in amplitude. The equipment, which was shown installed in a full-sized model of an aircraft, uses a 5-in cathode-ray tube and a scanner reflector about 2 ft across.

Also noted was a rear warning radar aerial in the tail unit of a French fighter aircraft (Fig 1).

Outstanding among the radio navigational aids on show was a new instrument landing system by the French company C.S.F. It has the advantage over previous systems of operating on a single frequency and of giving the facilities of azimuth, elevation and distance measurement to the pilot on dial-type instruments. As in other landing aids, overlapping radio beams with different modulation frequencies are used for azimuth and elevation measurement and an
interrogator-responder is used for the measurement of distance, the interrogation frequency being between 984 and $996 \mathrm{Mc} / \mathrm{s}$.

The use of a single frequency only is made possible by a form of time-sharing pulse transmission. The cycle of transmission is for an azimuth left-hand beam first, followed by an elevation lower beam and then an azimuth right-hand beam and finally an elevation upper beam. The pulse for each of these tranmissions lasts $1 / 65$ second and each is separated by an interval of $1 / 360$ second. There is then an interval of $1 / 40$ second, during which time the transmitter can send out the response signals to the airborne interrogator, using the distance-measuring facility. The cycle of operations is controlled by a rotary switch revolving 600 times per minute (Fig. 2) which feeds into waveguides and thence to the four aerials. Two of the aerials are for laying down the azimuth beams (known as the "localizer") and are placed upon either side of the runway. The other two are "cheese" aerials providing the elevation beams (known as the "glide-path"), one radiating the upper and the other the lower beam to form a $2 \frac{1}{4}^{\circ}$ descent guidance path. The pulses sent out are modulated with a square wave of $20 \mathrm{kc} / \mathrm{s}$ for the lefthand "localizer" lobe and $24 \mathrm{kc} / \mathrm{s}$ for the right-hand lobe. For the "glide-path" the upper lobe modulation is $34 \mathrm{kc} / \mathrm{s}$ and that of the lower lobe $30 \mathrm{kc} / \mathrm{s}$. The aerial for the distance measurement is a "cheese" with separate feeds for transmission and reception.
The airborne equipment consists of two small units and a power supply, together with a crossed-pointer "localizer" and "glide-path" indicator. Distance is indicated on an edge-scale instrument.
Signals can be received from the "localizer" transmitter at a distance of about 25 miles and the distance-measurement transmitter has a slightly shorter range. The "localizer" accuracy is to within $\pm \frac{12^{\circ}}{}$ at $1 \frac{1}{2}$ miles from the transmitter and the distance measuring facility has an accuracy to within 150yd up to 5 miles from the transmitter.
A new "talking beacon" made by the Swedish firm AB Gasaccumulator and intended mainly for fighter aircraft does not require the installation of any special receiver in the aircraft as it operates off the normal v.h.f. equipment on $100-150 \mathrm{Mc} / \mathrm{s}$.

The beacon depends upon sharply defined beams for its correct operation and these are obtained from two aerial arrays mounted back to back with a screen between them (Fig. 3). This permits forward and backward beams to be radiated without mutual interference. When, for instance, the forward beam is pointing in a northerly direction a transmission by voice of the course to steer to reach it is made, while at the same moment, when the backward beam is pointing south, a voice transmission of the reciprocal bearing is also made from a second transmitter. There is a limitation to the number of voice announcements which can be made, determined by the frequency with which information is needed in practice, and this has been settled at a repetition rate of every 30 seconds when flying on a constant compass heading. At this speed the announcements are made every 20 degrees of beam rotation. Since the aerials are relatively simple there are side lobes present which could give rise to false courses unless precautions were taken to prevent them. The side lobes are therefore masked by a third transmitter which is ten times more powerful than the voice transmitters and which feeds into an "H" type aerial
having a figure-of-eight polar diagram. This aerial is so placed that the wanted beams for the two bearings appear in the crevasses of the masking figure-ofeight polar diagram. Thus all of the side lobes are rendered inaudible by the modulation of the masking transmitter, which is synchronized in time with the voice announcements.

The "Narco Omnigator," made by the American National Aeronautical Corporation and exhibited by the French firm Air Tourist, was of considerable interest, being typical of American radio equipment technique for small aircraft. The instrument, which is compactly designed and only weighs 181 b , combines the functions of v.h.f. transmission on 8 channels, continuously tuned v.h.f. reception over the 108-127 $\mathrm{Mc} / \mathrm{s}$ band, instrument landing system (ILS), v.h.f. aural range (VAR), v.h.f. omni-directional range


Fig. 2. Rotary switch used at the transmitter of the C.S.F. instrument landing system.
Fig. 3. Swedish "talking beacon" showing the masking aerial mounted above the main array.



Fig. 4. Bendix radio compass with ferrite-rod aerial elements flush - mounted underneath an oircraft fuseloge.


Fig. 5. Cathoderay tube indicator used in the S.I.N.T.R.A.airborneradio compass.
(VOR) and " marker" reception. It is panel mounted and occupies a frontal area of only about $6 \frac{1}{2}$ in $\times 6 \frac{1}{2}$ in.
Flight demonstrations of the latest Decca Navigator flight $\log$ computers* were given throughout the period of the show. These devices, which operate from signals provided by Decca Navigator airborne receivers, are very much lighter in weight than previous computers and effect a saving of some 40 lb when used in a complete installation. They are entirely electro-mechanical, using no valves, and are designed for operating speeds up to 700 knots. To permit operation at this very high speed, a special memory circuit is included which prevents interference from "lane identification" signal breaks causing loss of "lanes" and is similarly effective during short signal interruptions.

In air navigation nowadays most of the official procedures for approach and departure to and from civil aerodromes are based on radio beacons and ranges. In some flying zones a rapid change-over from one beacon to another is required, and to enable this to be done as quickly as possible three of the French made radio compasses provide means for instantaneous change-over of frequency. "In the S.F.R. and Radio Air models a mechanical " memory" arrangement is included whilst in the S.I.N.T.R.A. equipment 18 pre-set, crystal-controlled frequencies are provided. In the mechanically tuned receivers the procedure for enabling the change-over to be made is for the pilot to tune the receiver to the second frequency required -such as the last beacon frequency wanted for a given procedure-and to close a "storing switch," after which he can use the radio compass on any other frequency until the second one is wanted, which he

[^3]simply obtains by opening the "storing switch." The time taken for the compass pointer to settle to the new bearing is between 5 and 8 seconds.
In two radio compasses exhibited, including the British Marconi (see last issue p. 306), fixed-coil loop aerials were used, while Bendix showed their new " magnetic antenna." This aerial (Fig. 4) does not entirely dispense with moving parts at the aerial proper but greatly reduces the size of the search coil, thereby increasing the rapidity of action of the direction finder when seeking a new bearing.

The theory of the device is based on the ability of high-permeability ferromagnetic materials to conduct magnetic lines of force easily and to draw into their conducting paths more lines of force than would be found in an equivalent area in free space. In the Bendix loop, four ferrite poles, each with a shortcircuited turn of wire round it, are placed at $90^{\circ}$ to one another around a small coil, and their effect is to increase the number of lines of force across the coil. The orientation of the lines of force across the coil depends on the relative signal strengths in the collector rods. Thus pick-up in the quadrants of the aerial causes the maximum concentration of energy across the collector rods which lie transversally to the direction of the transmitting station. The lines of force therefore travel across the loop coil in paths parallel to the collector rods which are receiving the maximum energy. At angles of reception lying between the quadrants, the relative signal amplitudes in the collector rods result in a shifting of the magnetic lines across the coil so that they are parallel to the field in free space.
In the radio compass developed by the French firm S.I.N.T.R.A. the indicator is in the form of a small cathode-ray tube with a graduated scale round its periphery (Fig. 5). This avoids any moving parts in the direction finder proper when used in conjunction with a crossed loop with fixed coils. Operationally the cathode-ray tube indicator has the advantage of immediate indication of the disappearance of the signal and of very rapid indication of transit over the top of a beacon or "coning."

When the loop is installed in the aircraft, one coil is arranged to point in the line of flight, leaving the other at right angles to it. When a signal is received in the line-of-flight loop, its amplitude is proportional to the cosine of the bearing angle of the transmitter being received relative to the heading of the aircraft, whilst the amplitude of the signal received in the other loop is proportional to the sine of the angle. Each loop is arranged to feed into an r.f. transformer and thence into the grids of a double triode valve which also have applied to them an a.f. modulation of equal amplitude but in phase opposition. The outputs from the anodes of the triodes are then fed into a balanced transformer, thereby suppressing the carrier, but leaving the sidebands. The r.f. signal is also received on a small vertical aerial and, after amplification, fed into the r.f. transformer, resulting in a carrier with two sidebands. The phase of the modulation of the carrier is a function of the bearing of the transmitter relative to the heading of the aircraft. The modulated carrier is then amplified and detected and, after suppression of the d.c. component, applied to a pulse generating circuit where a pulse is formed each time the detected signal passes through a maximum. This pulse is applied to the cathode-ray tube and appears as a notch in the circular sweep corresponding to the bearing of the transmitter in relation to the direction of flight.


Fig. 1. Coded information in the form of punched holes in paper tape being fed into a high-speed reading device.

AUTOMATION is the magic word that seems to have everyone by the ears (and possibly by the nose as well) in the mass-production industries just now. Radio and electronics engineers, however, appear to be taking the idea very calmly, even with indifference; seeing it perhaps as little more than a new term which has just been invented to describe techniques they have known about for years. And certainly if one looks through the radio and electronics literature for the past decade there is plenty of evidence to support this attitude.
Electronic techniques, of course, are not the whole of automation, but it seems they have a big part to play in the new sphere (in so far as anyone can say what it is). This came out fairly well at the recent conference at Margate on "The Automatic FactoryWhat Does It Mean?" organized by the Institution of Production Engineers. The titles of some of the papers were in themselves a fair indication-" Automatic Electronic Control of Machine Tools," "Atoms, Electrons and Automation," "The Computer-Electronics Contribution to Production," "Automatic Inspection-The Anatomy of Conscious Machines," "Computer-controlled Machine Tools," "The Automatic Office-Industry's Electronic Pulse." Nevertheless it was clear from the content of these papers that the advent of automation (or at least the talk of it) has not so far elicited anything new from electronics. It has merely taken a number of established techniques in electronic measurement, computing and control and placed them under the heading of Automation, together with other equally well-established techniques from mechanical engineering.

It seems, then, that the present conception of "The Automatic Factory" is little more than an ordinary factory to which a large number of electronic (and other) control devices have become attached like barnacles on a ship. This is not greatly significant to the electronics engineer, because it merely means an intensification of his work along the same lines and not a radically new approach. There is, however, a much more highly developed view of the automatic factory which sees the whole organization, and not just the individual bits of machinery, controlled and co-

## The

## Automatic

## WHAT SCOPE DOES IT

ordinated by automatic means. (It is sometimes forgotten that there is more to a factory than just the manufacturing processes themselves.) To use a biological illustration, whereas the former conception is equivalent to little more than a mass of disconnected local reflexes all working independently (such as the automatic control of the pupil of the eye with varying light), this highly developed idea is more like a complete animal, with all its sensing and actuating organs not only reacting to local conditions but controlled and co-ordinated by a central nervous system.

Here then is considerable scope for the electronics man-in the design of information-handling systems for overall control and in the linking of these systems to the local electronic control devices. To elaborate on the possibilities we cannot do better than quote from a recent edition of a book* by Norbert Wiener, the American mathematician, who is known principally for his writings on cybernetics. Describing an automobile factory of the future he says: "In the first place, the sequence of operations will be controlled by something like a modern high-speed computing machine. In this book and elsewhere I have often said that the high-speed computing machine is primarily a logical machine, which confronts different propositions with one another and draws some of their consequences. . . ."
"The instructions to such a machine, and here, too, I am speaking of present practice, are given by what we have called a taping (Fig. 1). The orders given the machine may be fed into it by a taping which is completely predetermined. It is also possible that the actual contingencies met in the performance of the machine may be handed over as a basis of further regulation to a new control tape constructed by the machine itself, or to a modification of the old one. . . ."
"The computing machine represents the centre of the automatic factory, but it will never be the whole factory. On the one hand, it receives its detailed instructions from elements of the nature of sense organs, such as photo-electric cells, condensers for the reading of the thickness of a web of paper, thermometers, hydrogen-ion-concentration meters, and the general run of apparatus now built by instrument companies for the manual control of industrial processes. These instruments are already built to report electrically at remote stations. All they need to enable them to introduce their information into an automatic highspeed computer is a reading apparatus which will translate position or scale into a pattern of consecutive digits. Such apparatus already exists, and offers no great difficulty, either of principle or of constructional detail. The sense-organ problem is now new, and it is already effectively solved.
"Besides these sense organs, the control system must contain effectors, or components which act on the outer world. Some of these are of a type already familiar, such as valve-turning motors, electric clutches and the like. Sonse of them will have to be invented,

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## Factory

OFFER ELECTROXICS?

to duplicate more nearly the functions of the human hand as supplemented by the human eye. . . ."
"Of course, we assume that the instruments which act as sense organs record not only the original state of the work but also the result of all the previous processes. Thus the machine may carry out feedback operations, either those of the simple type now so thoroughly understood, or those involving more complicated processes of discrimination, regulated by the central control as a logical or mathematical system. In other words, the all-over system will correspond to the complete animal with sense organs, effectors, and proprioceptors, and not, as in the ultra-rapid computing machine, to an isolated brain, dependent for its experiences and for its effectiveness on our intervention. . . ."
Undoubtedly Wiener's description is very much of a pipe-dream at the moment (or should we say a pipe-nightmare?), but it comes from an informed imagination and should not be dismissed too lightly. Already, in fact, we are beginning to see the initial developments. On the purely manufacturing side there are automatic electronic devices using "sense organs" and "effectors" coming into use for the control of continuous processes, while on the organizational side electronic digital computers, originally introduced into factories for straightforward accounting work, are being used to assemble data for production control purposes. What Wiener refers to as "the general run of apparatus now built by instru-
ment companies for the manual control of industrial processes" could be seen in great variety at the recent British Instrument Industries Exhibition. The newer "information-handling" side was also well represented. For example, both Fielden and Elliott were showing automatic devices for continuously monitoring industrial plant by sampling physical variables (e.g., temperature, flow, level, pressure) at various points and printing out the results on paper (Fig. 2). Electronic discriminating circuits detect when the values are above or below pre-set limits, and cause the appropriate alarms to be given. The British Iron and Steel Research Association were showing how the angular position taken up by a shaft in a selfbalancing servo system can be automatically registered in digital form as a decimal number and a coded version of it recorded as punched holes in a paper tape. Such "analogue-to-digital converters" are, of course, essential components in apparatus for numerical monitoring.

In another type of work an important development for the metal-working industries is the computercontrolled machine tool, which is now emerging from the laboratory and being sold as a commercial product. The Ferranti equipment, which has already been described in Wireless World $\dagger$, was the subject of a paper by D. T. N. Williamson at the Margate conference, and some of the computer circuitry was shown in a small exhibition which ran concurrently (see Fig. 3). The computer here is used for interpolation between points of change on the contour to be machined, and is a digital machine plotting out the curves point by point. R. H. Booth, of E.M.I. Engineering Development, showed, in another paper, how the same sort of interpolation could be achieved by analogue computing techniques, and there was

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Fig. 3. The digital computer used in the Ferranti computer-controlled machine tools is buitt up from replaceable plug-in drawers, each of which contains a number of logical units in the form of replaceable plug-in circuit cards. Most of the actual computing is done with semi-conductor diodes, and thermionic valves are used only for amplification between logical operations.


Fig. 4. Commercial electronic digital computer suitable for production planning. The unit on the left contains the programme controller and arithmetic unit, while on the right is a tabulating machine which feeds in information from punched cards (left-hand side) and automatically prints the computed results (right-hand side). This computer and the equipment in Fig. 2 were developed by the British Tabulating Machine Company.
considerable discussion on the relative merits of the two rival methods.

The use of computers for the overall planning of factory work was also discussed by Mr. Booth. He envisaged a machine into which could be fed information on the articles to be manufactured (in terms of the processes needed to make them) and out of which would be obtained information on such things as completion time, costs and requirements for fresh supplies of raw materials. To do this the computer would have to incorporate storage systems containing all the necessary reference data such as stocks of materials, available capacity for machining and assembly and costs of individual processes. A commercially built digital computer of the kind that could be adapted for such work was shown in the exhibition by the British Tabulating Machine Company (see Fig. 4). In the completely automatic factory, however, such a computer would do more than just produce results for human beings to act upon. Also fed into it would be the "operational" information from the plant monitoring systems described above, and this would be assimilated with the "planning" information to produce data from which the computer would control the plant automatically.
A more specialized type of computer for the control of continuous manufacturing processes was described by J. A. Sargrove and Peter Huggins, of Sargrove Electronics. It works on statistical principles and makes corrections to the processing machinery on the basis of error trends which it detects in the finished product. The apparatus (Fig. 5) includes a mechanical measuring device to sense the variations in the product, a transducer to turn these into electrical signals, a selecting device which samples the product at


# Wide Range Electrostatic Loudspeakers <br> 3-Complete Systems: <br> Loudspeaker/Room Relationships 

By P. J. WALKER*

|N the first part of this article we showed that for a given size, the apparent efficiency of an electrostatic unit may be increased by reducing the bandwidth which that unit is required to cover. An obvious method of increasing the overall efficiency of a complete electrostatic system, theretore, is to divide the system into a convenient number of frequency bands and to feed them via crossover networks. Optimum ciesign is obtained by increasing gaps and areas with decreasing frequency.

An alternative method of increasing apparent efficiency is to subdivide the loudspeaker area into a number of smaller units each covering the whole frequency range, the units being coupled by inductors so that the whole loudspeaker becomes a transmission line. (Fig. 1.) The acoustic radiation resistance appears as conductance in parallel with each capacitive element. For a fixed total area, and neglecting losses, the efficiency varies directly with the number of subdivisions.

Consideration of these two systems shows that frequency division has considerable advantages over transmission line divisions for most complete systems of domestic size and power requirements. First, if a


Fig. I. Capacitive loudspeaker elements coupled with inductances to form a transmission line.
single nine-octave unit is subdivided into a two-unit system, the apparent efficiency is increased 16 times. To obtain the same increase by transmission line division would require a minimum of 12 divisions. Unless the total area of the loudspeaker is large, and the plate separation small, the capacitance of each section of the transmission line becomes very small indeed and requires correspondingly large inductance which must be of relatively high $Q$.

This apparent efficiency advantage of frequency subdividing over transmission line dividing holds until the bandwidth of each unit is reduced to two octaves.

Apart from transmission line subdivision applied to individual units of a frequency-divided system, practical consideration normally limits transmission line techniques to large-area diaphragms. When such is the case, however, additional facilities are available to the designer both in the accurate control of directional characteristics and in providing a constant phase contour, independent of frequency.

In discussing various possible forms of complete

[^6]electrostatic systems, a novel situation arises. The quality criterion of a loudspeaker usually concentrates on three performance parameters, as measured in an unlimited atmosphere. (a) Ability to produce a required sound intensity over the audio spect-um with negligible non-linearity distortion. (b) The sound pressure over the designated listening area should be independent of frequency throughout the audio range. (c) Operation should be aperiodic.

Complete loudspeakers designed on the principles which we have been discussing are capable of meeting these three requirements to a new and exciting degree. We shall see that different designs and approaches differ not so much in terms of (a), (b) and (c) above, but in other factors of importance to quality reproduction; factors which have previously had to take second place or have been masked in the struggle for (a), (b) and (c).

## Corner Mounting

There has been a strong tendency in loudspeaker design to make use of the corner of a room. This is because at low frequencies the air load resistance for a given size of diaphragm is increased 8 times over that of an unlimited atmosphere.

Since the ratio of cabinet "stiffness" to air load resistance is independent of diaphragm size, any increase of resistance due to boundary walls and floors fundamentally reduces the size of cabinet required for a given performance.

As an example, the form of corner electrostatic loudspeaker illustrated in Fig. 2 and designed for full performance down to $40 \mathrm{c} / \mathrm{s}$ utilized an internal resonance with a $Q$ of 3 and a built-in enclosure of 10 cu ft . Fundamentally the enclosure size could be reduced either by (1) increasing $Q$, (2) reducing power and apparent efficiency requirements, or (3)

Fig. 2. Wide-range electrostatic loudspeaker in a resonant corner enclosure.


Fig. 3. Cylindricol electrostatic loudspeaker. Each strip corries the full frequency range and the sections are coupled to form on electrical tronsmission line. The inductor assembly is shown below.

restricting frequency range. Any one factor may be traded for any or all of the others.

It should be noted that with the diaphragm area of Fig. 2 the resistance could be substantially increased by reshaping the whole of the low-frequency area near the floor so that, with the boundary reflections, its dimensions laterally and vertically are similar. Such a form, with a suitably shaped treble unit above it, can be designed to give a level response in direct radiation to the listening area, so that the (a), (b), (c) requirements are not affected. Homogeneity on the other hand, due to the physical spacing of units, is destroyed. This may be more important than is generally realized, particularly in rooms of normal domestic size.

The high-frequency section (centre strip in Fig. 2) is sealed at the rear by an enclosure of width equal to that of the strip and incorporating a fibreglass wedge to offer almost pure resistance throughout the range of the unit. This sealing is necessary in order to maintain front air load resistance by preventing coupling between front and back.

Fig. 3 shows an entirely different form of corner design. The diaphragm area covers the whole surface and extends around the back to form an enclosed cylinder. Every part of the diaphragm carries the whole frequency range. The surface area is divided into units to form a transmission line. The total volume is 15 cu ft . The step in diameter is introduced because the transmission line rotates around the top portion and thence around the bottom portion. The time delay in the sound expanding from the top portion to the diameter of the bottom portion is equal to the time delay of the electrical voltages in the transmission line.

The complete assembly is placed a small distance from the corner of a room so that the boundary reflec-
tors are aiding at the lowest frequency of interest. The large diaphragm area together with the boundary reflections provide a loading approximately equal to $\rho \mathrm{c}$ at $30-40 \mathrm{c} / \mathrm{s}$. Internally there is acoustic resistance treatment, so that there will be resistive loading at high frequencies, changing to a capacitive load due to the lumped enclosure at low frequencies. Simplified equivalent circuits for high and low frequencies are shown in Fig. 4. The turnover occurs at about $400 \mathrm{c} / \mathrm{s}$ and it is obvious that with constant voltage the response will be level above $400 \mathrm{c} / \mathrm{s}$ and drop at $6 \mathrm{~dB} /$ octave below this frequency. This is corrected by progressively rematching to the amplifier below $400 \mathrm{c} / \mathrm{s}$. The section shape may be elliptical to give a degree of direction at high frequencies.

It is obvious that the corner boundaries will introduce peaks and troughs throughout the frequency range. These are, however, exactly the same as occur naturally with live speech or music originating near boundaries in a room. To what degree these effects are important must at the present time be a matter of conjecture. It can safely be said that the subjective effect is by no means as alarming as the appearance of the response curve.

The advantage of a corner position has already been noted. This advantage is not gained without considerable detriment in other directions. If we wished to excite every room resonance to its fullest extent with a sound source of high internal impedance, we put this source in a corner because this is the position of highest impedance for every mode. In placing our loudspeaker in a corner therefore we are placing it in the worst possible position if our aim is smooth aperiodic sound.

Although the present trend appears to be to tolerate this state of affairs in the interest of the organ's 32 ft rank (or reduction of cabinet size), the inherent smoothness of electrostatic loudspeakers once experienced is not lightly thrown away, and there is added impetus in attempts to improve the loudspeaker/room relationship.

## Double Wall Enclosure

The strip "twin" unit design of Fig. 2 may be built into a wall in such a way that most room modes are not excited or are excited only feebly. If it is an outside wall, the rear enclosure may be added externally. If an inside wall it may spread over the wall so that from the appearance point of view it has virtually disappeared. Fig. 5 shows the general form of installation. The strip unit extends from floor to ceiling and the low-frequency sections are backed by 5 in wide enclosures $4 \frac{1}{2} \mathrm{ft}$ in length, with fibreglass wedges incorporated. The impedances and response are shown in Fig. 5 (June issue). With the dimensions of this example, $d=10$ in since both 5 in units are coupled, and the response will be within 3 dB of $1 \mathrm{kc} / \mathrm{s}$ response down to $35 \mathrm{c} / \mathrm{s}$. These figures include floor, one wall and ceiling, but do not, of course, include the effects of other room boundaries. Assuming a 2 in thick wall for rigidity, the volume of a room of 300 sq ft floor area would be reduced by $2 \%$.

There can be no initial excitation of floor to ceiling modes because vertical excitation is evenly distributed. Modes excited in a direction parallel to the wall on which the speaker is mounted will be reduced in number. Assuming a rectangular room, the number of modes excited will be some four times less than the number excited by a corner floor position.

As can be seen by the following summary, this form of loudspeaker leaves little to be desired.

1. The enclosure being "built-in" can be completely rigid.
2. The only fold in the enclosure is narrow compared to wavelength and being close to the diaphragm can cause no reflections in the range of that unit.
3. The loudspeaker and its enclosure are completely predictable.
4. The (a), (b) and (c) requirements previously mentioned can be met virtually to perfection.
5. Radiation throughout the whole frequency range is homogeneous; there is no source displacement and no phase problems at crossover.
6. Total radiated energy (as well as axial pressure) is independent of frequency.
7. The loudspeaker/room relationship is good.

Item 6 deserves further mention. The normal frequency response specification of a loudspeaker is in terms of sound pressure produced on the axis or over a limited listening arc. The mean spherical radiation (total power output) is not usually specified, although it will have a profound effect in a room because the intensity of indirect sound is dependent upon it. If high-frequency radiation is limited to a segment of, say, $90^{\circ} \times 30^{\circ}$ (a typical figure) and bass radiation is hemispherical, and if the axis response is level, then there will be a step of 12 dB in the mean radiated response. This produces an artificial step in the acoustic ratio (ratio of direct to indirect sound) producing unnatural hardening of the reproduced sound.

## Doublet Sources

We now come to consideration of the doublet as a sound source and we shall see that it possesses properties of considerable significance in improving loudspeaker/room relationships. By a doublet we mean a diaphragm, radiating on both sides.

If we assume a $12 \mathrm{in}-15 \mathrm{in}$ unit (moving coil or electrostatic) mounted in a 4 ft - 5 ft baffle, we find that the


Fig. 4. Equivalent circuits at high and low frequencies of the acoustic loading on the loudspeaker of Fig. 3.


Fig. 5. Sectional plon showing one method of rear enclosure for a strip electrostatic unit.
acoustic system has three main faults. (1) The acoustic air load falls to very low values at wavelengths larger than the baffle size. (2) The acoustic load is very irregular at low frequencies and (3) reflections from the baffle edge occur at higher frequencies. The second, and third faults can be mitigated by adopting peculiar shapes.

If, instead of a baffle, we construct a composite electrostatic unit of the same area, the position is completely altered. The resistance per unit area and the total working area are both increased so that the air load is many times that of the baffle case. The load, and consequently the performance, is regular and predictable.

The construction is that of strip units progressively increasing in plate spacing and area from the centre line. Due to the air load resistances involved for each strip, the permissible bandwidth is reduced over that which could be obtained if the back radiation were sealed off and it is necessary to split the frequency range into three to obtain efficiency comparable to a two-way " sealed" system.

Any unloaded strip considered alone will have a resonant frequency when the diaphragm stiffness reactance equals the air load mass reactance. This is, however, placed below the frequency range of the strip, so that the mutual radiation of the adjacent strip carrying a lower frequency range increases the radiating area and prevents the application of any effective mass. The complete system is therefore entirely free of resonance except at one low frequency (usually placed at $30-35 \mathrm{c} / \mathrm{s}$ ). The Q of this resonance is adjusted to mantain response to this frequency.

The complete loudspeaker has a cosine characteristic and this is substantially maintained through the range. It cannot radiate sound in the direction of its surface, horizontally or vertically, so that it cannot excite room modes in two out of the three room dimensions. It will only excite modes in the remaining dimension when placed at a region of maximum velocity for that mode. (The impedance looking into the loudspeaker is low.)

Having a "cosine" polar characteristic the mean spherical radiation is reduced by a factor of 3 at all frequencies, so that quite apart from freedom of mode excitations any colour due to the room is reduced by a factor of three. This is exactly analogous to a "velocity" microphone. In the same way that a "velo-" city" microphone is used in place of a "pressure" microphone to reduce studio colour, this "velocity" speaker will reduce colour due to the listening room.

Listening tests comparing "pressure" and "velocity" speakers of otherwise similar characteristics indicate that a velocity characteristic may well have important features for high-quality reproduction. An electrostatic loudspeaker of this type correctly positioned in the room meets all requirements as did the "wall" form previously described, with the addition of an even better loudspeaker/room relationship. The fact that it requires to be free standing well within the room may or may not be advantageous.

The more the acoustic ratio is reduced (provided always that it is reduced equally at all frequencies), the more one approaches the state of affairs that the pressure at the ears is a replica of the pressure at the position of the microphone in the concert hall or studio (ideal headphone conditions). It must be emphasized that many arguments for and against this condition have been proposed. It is outside the scope


Fig. 6. Stereophony from a single transmission line loudspeaker, with separate channels feeding each end of the line.
of these articles to enter these arguments other than to say that with a monaural channel the choice must be an æsthetic one.

A complete listening room can be designed to produce pressures throughout the room which are more or less equal to the pressures at the studio microphone.

A tube of small diameter compared to wavelength fitted with a piston at one end, and terminated at the other by a resistance of $\rho c$ will give pressures anywhere in the tube which are directly proportional to piston velocity and independent of frequency. Provided that the area of the piston equals the tube crosssectional area, then the requirement of small diameter disappears.
A rectangular room with a diaphragm covering one wall and correct termination on the opposite wall meets the requirements. The space behind the diaphragm must be at least l0in deep and treated like the speaker in Fig. 3. The equivalent circuit is the same as Fig. 4. The sound absorption treatment of the opposite wall must ideally be several feet in depth.
Sound intensity throughout the room is independent of position (including the distance from the diaghragm). The apparent sound source is always in a direction perpendicular to the diaphragm and, of course, moves as the listener moves.

The same loudspeaker may be used for stereophony. With transmission line matching and feeding the signal at one end the wavefront will be tilted, due to time delay. Separate signals may be fed from either end to produce two tilted wavefronts, one for each signal. Since each apparent origin is perpendicular to its wavefront, the aspect angle from the listener is a constant and entirely independent of the listener's position over a large triangular area (Fig. 6). The relative intensity of the two signals is also constant.

A fixed angle, two-channel system of this type may be obtained with a less elaborate listening room. The strip arrangement of Fig. 5 may be installed horizontally instead of vertically. If each unit is a transmission line along its length, then two cylindrical wavefronts will be produced with exactly the
movement with frequency (b) Homogeneity, (c) Acoustic ratio, (d) Mode excitation, (e) Phase contour, etc. All are factors which can only be tentatively assessed after long usage.

The author wishes to thank Ferranti Ltd. for permission to publish the result of work jointly carried out, and to acknowledge the invaluable work of D. T. N. Williamson, W. D. Ol iphant, and their colleagues at Edinburgh. Thanks are due to J. Watson, J. Collinson and others at the Acoustical Manufacturing Company, and to the several specialists who have been able to assist with problems in their own field.

## COMMERCIAL LITERATURE

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Multi-range Test Meter covering a.c. and d.c. voltage, each in six ranges, a.c. current in one range, d.c. current in six ranges and resistance in two ranges. Specification on a leaflet from Measuring Instruments (Pullin), Electrin Works, Winchester Street, Acton, London, W.3.

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Band-III Television Convertor with PCC84 s.f. amplifier and ECC81 oscillator and frequency changer. Models available for double-sideband and single-sideband receivers (for London area). Brief techrical specification on a leaflet from Invicta Radio, 100, Great Portland Street, London, W.1.
Tape-to-Disc Transfer Service and other recording services Leaflet giving brief details and prices of records from Sound News Productions, 59, Bryanston Street, Marble Arch, London, W.1.

General view of the complete fitment. At the left-hand side of the centre section is the $15 i n$ speaker, with the treble speaker assembly on the top of the fitment. The only part of the top that opens is on the extreme right-hand side where the transcription motor is situated. The upper part of the centre of the fitment is a steel drawer in which are contained the tape recorder, f.m. tuner, preamplifer, amplifier and crossover unit. Three drawers underneath each accommodate 40 boxes of tape. Five cupboards provide storage space for records or tape and in nine small drawers tools and stationery are kept. The drawer underneath the gramophone turntable was made to correct dimensions to house a ribbon microphone.


# Migh-Fidelity Honse 

By Richard Arbib

## A sOLUTION TO THE PROBLEM OF LIVING WITH ALDIO EQUIPMENT

I$T$ is hard to determine whether it was the difficulty of finding room for the 91 st tape, the wish to have sand in the sitting room, or the fact that the new Garrard 301 transcription motor was so massive that it would not fit along the side of the Leak " Varislope $2 "$ pre-amplifier in the only available space in the existing cabinet. They had both been mounted temporarily in tiers and it was very awkward to adjust the controls on the pre-amplifier by threading the hand under the transcription motor. Incidentally, if it is thought that the possession of 91 tapes is plutocratic, it should be remembered that quite a number of people collect thousands of stamps.

Built-in Furniture Fitment. At last the decision was made. The window alcove in the sitting room should house an elaborate furniture unit to incorporate all the equipment which had been spread around the room or kept in the radio-gramophone cabinet. This had been made specially, more than a quarter of a century ago, with the idea that it would accommodate any developments of the future. However, this thought had now been proved to be erroneous, for it was designed at a time when life was simple and gramophone records were made to play at only one speed, tape machines had not been invented and only one loudspeaker was used at a time in a room.

In the window alcove there had been a writing desk, and at once an argument arose as to how one was to write letters without a desk. However, it was remarked that in this electronic age, one should not write letters at all, and writing was needed only for cheques. Any correspondence could be undertaken on tape and either transcribed by a secretary or heard by the recipient.

As is usually the case in planning any new development, the first thing was to undertake some research
and thus many Audio and High Fidelity American magazines were perused. In their pages are illustrated the elaborate high-fidelity installations which, with ranch houses, Cadillacs and typewriters that type as though they are printing, have become the essential acquisitions of the successful American executive. However, an examination of these designs showed clearly that whilst they housed all the pieces of equipment which were considered necessary for the addict of high fidelity, much of the apparatus was situated in a way which was most difficult to operate. It is very nice to have a tape recorder built into a room, but many will agree that it is inconvenient to have to lie on the floor to change the reels.

Basic Principles. In designing this equipment, therefore, it was decided that three basic principles had to be observed. First, everything had to be right technically. For example, the pickup must be some distance away from the loudspeaker. All equipment must be accessible for operation and must be capable of being withdrawn from a fitment in a time not exceeding five minutes. To the service engineer who has to struggle with the servicing of many television sets in a day, this time may appear to be excessive, but in view of the reliability of the units concerned, it was considered that five minutes would not be an undue time. Furthermore, it was also thought that every standard unit should be unaltered. For example, the tape recorder should not be taken out of its original case, for if any section did want servicing, it could be withdrawn as a complete unit and thus returned easily to the manufacturer.

After having attempted for some years to read the titles on tape boxes sideways in bookcases, it was decided that the majority of the tape boxes should be kept in drawers in such a way that they could be located easily. It was thought that the basic storage
space available should be capable of housing 10012 in records and 200 reels of tape.

Although the window alcove was 7 ft wide and 5 ft deep, one snag was that a height of only 31 in was available under the windows. However, it was agreed that the whole of this space should be occupied by the units, loudspeakers and nine small drawers for stationery and odd tools which had been kept previously in the desk. As, apparently, most wives insist on putting vases of flowers, photographs, ash trays and other bric-à-brac on top of furniture, it was arranged that only one small portion of the installation should have a lid which opened. Although record changers may be placed in drawers, it was realized that a transcription motor with a pickup weighing only a few grams should not be mounted in anything movable, and thus of the whole surface area of 22 sq ft only 3 sq ft lifts up.

Accommodating the Units. The basic units which had to be accommodated comprised a 15 -in Wharfedale loudspeaker, which, from the counsels of Mr. Briggs, it was appreciated must be housed in a cabinet exceeding $9 \mathrm{cu} \mathrm{ft}, 8$-in and 5 -in speakers which should be mounted in an open form of baffle, a Ferrograph Model 2A tape recorder, a Leak TL12 amplifier and "Varislope 2 " pre-amplifier, an American Browning f.m./a.m. tuner, which had been brought over to this country long before British f.m. tuners were available, and the Wharfedale 3-way cross-over network unit. Furthermore, a panel was required for switches. The gramophone equipment, for which space had to be found, was a Garrard 301 transcription motor with a Leak moving-coil pickup and arm. Arrangements had to be made for feeding the output to the loudspeaker network, which has been in existence for some time and extends to practically every room in the house.

Owing to the comparatively low height of the top level of the fitment, it was decided that the only way in which to operate the Ferrograph and the amplifier controls conveniently, was to mount everything, except the loudspeakers and the transcription motor, in a drawer. From the accompanying illustrations it will be clear how the units are arranged. The steel drawer was made in two sections and welded together, the Ferrograph drops in on the left-hand side, the TL12 amplifier is in the base of the right-hand side. Mounted above it on a shelf is the Wharfedale crossover network unit to the controls of which extension rods have been fitted so that the potentiometers on the medium and treble cross-over network sections are controlled by knobs mounted on the top panel. The f.m. tuner unit is mounted behind and the little available space in front of that is occupied by a felt-lined recess in which a small crystal microphone is kept and which is connected permanently to the appropriate socket on the Leak pre-amplifier. This unit is mounted at the front of the drawer at the same angle as the control panel of the Ferrograph. A ribbon microphone is used also and this is kept in a special drawer below the gramophone motor.

A potentiometer has been added at the tape input socket of the Leak "Varislope 2 " so that the levels from the Ferrograph, the tuner unit and the pickup are all the same. On the left of the "Varislope 2 " is a special control panel which has four switches and four jack sockets. Reading from left to right (1) switches the TL12 amplifier to the three loudspeakers through the cross-over unit, (2) switches the amplifier to the extension loudspeaker network, (3) switches the
output of the Leak amplifier, either to the three loudspeakers through the cross-over unit or direct to the 8 -in loudspeaker, when it is required to reproduce speech only, and (4) switches off everything at the mains, irrespective of the positions of the switches on the individual units. The jacks below are: (1) connected to the three loudspeakers, (2) connected to the external loudspeaker network, (3) connected to the output of the tuner, and (4), connected to the sound output of the television set which is situated in another part of the same room. The jack sockets are provided for the three loudspeakers and the extension network in case it is wished to feed the output of the Ferrograph direct to these speakers without going through the Leak amplifier. In the same way the tuner socket is used for feeding from the tuner into the Ferrograph. The tuner is also connected permanently to the appropriate socket on the pre-amplifier.

Operation.-One length of shielded cable with screened jack plugs at each end and the various sockets are used as follows. If it is wished to record a radio programme the output is taken from the tuner socket on the small panel and fed into Input 2 of the Ferrograph, the monitoring to the tape machine being undertaken by the gain control on it. A programme can be recorded without it being heard in the room. or if it is wished to be heard in the same room or another room the appropriate loudspeaker switches are thrown and the gain control on the "Varislope 2" adjusted to a suitable volume. If it is desired to record a television programme, the jack is plugged into the TV socket on the sub-panel. If the TV programme is desired to be heard through the Leak amplifier either in the same room or around the house, the output is fed from the TV socket to the input jack socket on the pre-amplifier. To record speech from the room the main control switch on the preamplifier is turned to "microphone" and the output from the pre-amplifier connected to Input 2 of the Ferrograph. If it is desired to reproduce a gramophone record on tape, the same procedure is observed with the Leak main control knob in the position to suit the frequency response of the record concerned.

Loudspeaker Mounting.-The arrangement for the loudspeakers is that the section of the fitment for the 15 -in speaker has the sides made of 1 -in wood, 1 -in of sand and $\frac{1}{2}$-in of plywood. The front has a port 12 -in $\times$ $6-\mathrm{in}$. On the main top of the fitment is situated the assembly for the 8 -in and 5 -in speakers. As the height of the alcove is only 7 ft the two speakers do not point upwards but are at an angle of $45^{\circ}$ into the room. In order to obtain a simple and attractive design, they are mounted on a flat baffle which is held by skeleton woodwork at an angle of $45^{\circ}$, the front and the sides being covered by expanded metal, and the back is open.

The Garrard 301 transcription motor is mounted on a motor board which floats on springs. Also mounted on the motor board is a light which illuminates the turntable and an extra switch which stops the motor without any danger of the pickup being moved.

Cabinet Work.-The finish of the whole cabinet work is bird's-eye maple for the exterior of the front and interior of the cupboard doors, motorboard and panel above the tuner unit. The top is of walnut and the drawers are made of mahogany veneered on the outside with bird's-eye maple.

The front of the steel drawer which accommodates the recorder, amplifier, f.m. tuner, etc., is a dummy
to the extent that it has five handles or knobs but opens as one drawer. The units in this drawer have, with the drawer, a combined weight of about 2 cwt . The drawer is mounted on Admiralty pattern ball-bearing slides and moves forwards and backwards easily. The wood front is hinged at the top so that the recorder and pre-amplifier controls may be operated without pulling out the drawer.

Having all these units in a comparatively confined space does, of course, raise the problem of ventilation. This has been solved by having the underside of the left-hard section of the drawer completely open except for two channels on which the Ferrograph rubber feet rest. The sides of this section of the drawer are cut away to coincide with the loudspeaker fret and ventilation holes of the recorder case. The other side of the steel drawer has slots $17 \mathrm{in} \times 2$ in cut at the lower ends of the back and sides. Holes are provided at the top of one side, in the felt-lined recess, and an expanded metal grille is mounted between the preamplifier control panel and the switch-jack panel, which have purposely been fixed half-an-inch apart.

The large drawer at the side and the two underneath the steel drawer are of the exact dimensions to take boxes of 7 in and $8 \frac{1}{4}$ in reels of tape respectively. Each drawer takes two rows of 20 tapes which are filed with the titles uppermost and in line with the front of the drawer. Three shelves in the large right-hand cupboard each accommodate 25 tapes. Four smaller cupboards, whilst being of the correct dimensions to accommodate 12 in records could, of course, be used for the storage of tape.

Having spent many hours in making the necessary drawings for the furniture fitment and undertaking all the wiring and connections, of course the question asked most frequently by other high-fidelity enthusiasts is, what alterations would be made if it was being designed all over again. The answer, quite honestly, is none. There is only one snag to the whole system which was apparent when it was planned originally, and that is the desirability of having the main loudspeaker assembly on the side of the room opposite the
control equipment, so that the volume could be adjusted from a distance. With the space available this was not possible, and the only refinement which could be introduced to the system other than stereophonic sound, for which space is available, would be remote operation for the main gain control of the amplifier.
What Next?-Other friends have said, "Well, now you have everything, what can you do in the future, other than record and listen?" Apart from stereophonic sound which will, of course, be coming along later this year, work has already started upon rewiring the rest of the house with a system whereby the loudspeakers in each room will have individual volume controls and be operated through fourway switches. By this system the occupants of each room and the garden will have choice of the tape or record programmes from the equipment in the sitting room or the Home, Light or Third B.B.C. f.m. programmes. This system is a development of an idea suggested by Douglas Lyons, of the Trix Electrical Co., Ltd. In one cabinet in another room three f.m. tuners, each tuned to a different programme, will be connected to three 10 -watt amplifiers. The outputs of these amplifiers will be connected to three pairs of the four-pair cable. The remaining pair replacing the existing extension speaker pair on the main sitting room equipment.

The three f.m. tuners and amplifiers will be switched on and off by a time clock set to switch on at 7 a.m. and off at midnight.

Acknowledgements. - In conclusion, grateful acknowledgement must be made for the assistance of the following friends who have given much technical advice which has been the means of avoiding many of the pitfalls which fall to the experimenter who tries to put together the parts of different manufacturers. Reference is made particularly to Hector Slade of Garrard Engineering \& Manufacturing Co., Ltd., Harold Leak of H. J. Leak \& Co., Gilbert Briggs of Wharfedale Wireless Works, Ltd., and E. H. Niblett of Wright \& Weaire, Ltd. Yes, this article was dictated on tape!

The steel drawer pulled out with the flap covering the controls turned down. The tape recorder is on the left with a tape splicer mounted below the main control knob. Set at the same angle as the recorder control panel is the switch panel and pre-amplifier. On the panel above are the knobs controlling the potentiometers of the cross-over unit and the dial and controls of the f.m. tuner. in the felt-lined recess a crystal microphone is kept and is permonently connected to the preamplifier. The flap at the top right-hand corner is up to disclose the transcription motor and moving coil pickup. A lamp to illuminate the turntable is fitted on the motor board.


# Transistor Equivalent Circuits 

2.-Earthed-Emitter function Transistors

By W. T. COCKING, M.I.E.E.

IN Part 1, we developed an equivalent circuit for the thermionic valve. In doing this we started with the experimental evidence of the characteristics of the valve in the form of a family of curves relating anode voltage to anode current for a series of values of grid voltage. We then approximated these accurate curves by a series of equally-spaced parallel straight lines, which we may conveniently call a linear approximation. The next step was to find an equation which would represent this approximation algebraically and, finally, we found an arrangement of components which could be represented by the same equation. This formed an equivalent circuit.

The accuracy with which an equivalent circuit represents a valve depends entirely upon the goodness of the linear approximation. Over any range of voltages and currents where the straight lines fit the valve curves closely, the accuracy is good.

With the transistor, the procedure is exactly the same as with the valve. There is a complication, however. It is an experimental fact that all the electrodes of a transistor pass current, not two of them only as in the negative-grid valve. The transistor is more like a positive-grid valve in this respect. Because of this, two families of characteristic curves are needed to describe it and the equivalent circuit is more complex than that of the valve.

The newcomer to the transistor is apt to be misled by certain conventions which are commonly adopted. Most published characteristics are for the transistor in the earthed-base condition and many circuit diagrams show it used in this way. Also, in the usual transistor symbol, the base is represented by a heavy line which is not unlike the cathode symbol of a valve. It is hardly surprising, therefore, that the newcomer gets the impression that the base of a transistor corresponds to the cathode of a valve.

The base is, of course, the equivalent of the grid and the earthed-base transistor is analogous to the earthed-grid valve circuit. To anyone accustomed to valve circuits, it seems very wrong-headed to make the earthed-base circuit the basic one of transistor circuit theory. Actually, there are reasons why it is more suitable than the earthed-emitter circuit, but this is not the stage at which they can well be appreciated.

To the newcomer, however, the earthed-emitter circuit is the natural approach and this is the one that we shall adopt here. At first, only this circuit will be considered and only the junction transistor. This exists in two forms, the $n-p-n$ and the $p-n-p$. It consists essentially of two pieces of $n$ or $p$ type impurity germanium with a thin slice of the opposite kind sandwiched between them.

One of the end pieces functions as a source of

SUMMARY: As a sequel to the derivation of the equivalent circuit of the thermionic valve in Part 1, equivalent circuits for the $n-p-n$ transistor in the earthedemitter circuit are developed here. They are based upon the same form of straight-line approximation to the transistor characteristics as was used for the valve. The characteristics of the transistor are represented by four quantities which are easily derived from the characteristic curves; they are base and collector a.c. resistances and forward and backward current-amplification factors.
It is shown that, within the limits of accuracy imposed by the straight-line representation of valve and transistor characteristics, the transistor is exactly equivalent to a thermionic valve having a cathode feedback resistance and a second resistance between grid and cathode.
charge carriers and the other as a collector of them; the one is called the emitter and the other the collector. The meat in the sandwich functions as a control electrode and is called the base. Emitter, base and collector are analogous to the cathode, grid and anode of a thermionic valve. With the $n-p-n$ transistor the analogy is rather close, because the internal conduction is mainly by electrons, with the result that base and collector are normally maintained positive to the emitter and the conventional currents enter at base and collector and leave at the emitter.

The $p-n-p$ transistor, however, is rather different. Internal conduction is by positive "holes." Base and collector must normally be negative to the emitter and the current enters at the emitter and leaves by the base and collector. There is no valve which behaves like this but if it were possible to make one with a "cathode" which emitted positrons instead of electrons it would do so.

The circuits of typical $n-p-n$ and $p-n-p$ transistors in the earthed-emitter connection are shown at (a) and (b) respectively of Fig. 1.

## The n-p-n Transistor

In Fig. 2 are shown the general forms of the characteristics of an $n-p-n$ transistor. The family of curves at (a) relates collector voltage and current for a series of values of base current, while the family at (b) relates base current and voltage for a series of values of collector current. If the curves at (a) were for values of base voltage $V_{b}$ instead of base current $I_{b}$, they would be of exactly the same form as an ordinary set of valve curves and the ordinary equivalent circuit could be used.

Transistor characteristics can be plotted in terms of $\mathrm{V}_{b}$, but it is customary to use $\mathrm{I}_{b}$ instead, mainly because a transistor has such a low input resistance that we are often more interested in current than in voltage.

In Fig. 2 (a), the dotted lines represent a linear approximation to the real characteristics upon which we base our equivalent circuit. We proceed in the

Fig. 1. The circuit diagrams of a simple transistor amplifier are shown at (a) for an $n-p-n$ and at (b) for a p-n-p transistor.

(a)

(b)
same way as in the case of the valve, which was described in detail in Part 1. Some line, not shown, passing through the origin, represents a resistance of value $\rho_{22}=\mathrm{V}_{d} / \mathrm{I}_{c}$. The slope resistance $\delta \mathrm{V}_{c} / \delta \mathrm{I}_{c}$ has the same value for this line. For any other line the value $\mathrm{V}_{c} / \mathrm{I}_{c}$ does not apply but, since the lines are all parallel. they all have the same slope resistance which is, therefore, defined as

$$
\begin{equation*}
\rho_{22}=\delta \mathbf{V}_{c} / \delta \mathbf{I}_{c} \tag{1}
\end{equation*}
$$

We are, for the present, using the Greek letter rho to represent resistance instead of the more usual $r$, because the latter is commonly used in equivalent circuits derived for the earthed-base connection and we want to avoid any confusion between the two. We are using the subscripts " 22 " to denote what, by analogy with the valve, we may call the collector a.c. resistance, because that is customary with transistors. This resistance, which we call $\rho_{22}$, is the transistor equivalent of the anode a.c. resistance of a valve.

The particular line in Fig. 2(a) for zero base current

(a)

(b)

Fig. 2. Collector (a) and base (b) characteristic curves fol an $n-p-n$ transistor. The dotted lines represent ideal straight-line approximation to the characteristics.


Fig. 3. D.C. equivalent circuit of an $n-p-n$ transistor. The output side is shown at (a) and the input at (b).
( $\mathbf{I}_{b}=0$ ) cuts the current axis at some current $\mathrm{I}^{\prime}{ }^{\prime}$. The equation for the line through the origin is

$$
\mathrm{I}_{c}=\mathrm{V}_{c} / \rho_{22}
$$

The equation for the line $\mathrm{I}_{b}=0$ becomes simply

$$
\mathbf{I}_{c}=\mathbf{I}_{c}^{\prime}+\mathbf{V}_{c} / \rho_{22}
$$

and this derivation is exactly analogous to that for a pentode valve, Fig. 9 of Part 1.

We now define a current amplification factor

$$
\begin{equation*}
a=\delta \mathbf{I}_{c} \delta \delta \mathbf{I}_{b} \ldots \tag{2}
\end{equation*}
$$

for constant $V_{c}$. This is merely the ratio of a change of collector current to the change of base current which causes it. The complete equation for Fig. 2(a) now becomes

$$
\mathbf{I}_{c}=\mathbf{I}_{\underline{c}}+a \mathbf{I}_{b}+\mathrm{V}_{\mathrm{c}} / \rho_{22}
$$

which is more conveniently written as

$$
\mathrm{V}_{c}=\mathrm{I}_{c} \rho_{22}-\mathbf{I}_{c}^{\prime} \rho_{22}-a \mathbf{I}_{b} \rho_{22}
$$

Now what does this represent? On the left-hand side, $\mathrm{V}_{c}$ is the externally applied collector-emitter voltage and is not part of the equivalent circuit of the transistor itself. $I_{e} \rho_{22}$ is an internal voltage drop due to the current $\mathrm{I}_{c}{ }_{c}$ flowing through a resistance $\rho_{22} . \quad \mathrm{I}^{\prime}{ }_{c} \rho_{22}$ and $a \mathrm{I}_{b} \rho_{22}$ can be taken to represent e.m.fs acting round the circuit to assist $V_{c}$ but internal to the transistor. The first term of the pair accounts for the offsetting of the $I_{b}=0$ line of Fig. 2(a) from the origin; the second accounts for the effect of base current upon collector current.

The complete circuit which the equation represents is thus one like Fig. 3(a) in which the transistor part is shown boxed. It is the same as a valve equivalent save for the labelling of the elements.

Now consider the base characteristics of Fig. 2(b). A line through the origin represents a resistance $\rho_{11}=\mathrm{V}_{b} / \mathrm{I}_{b}$. This has a slope resistance $\delta \mathrm{V}_{b} / \delta \mathrm{I}_{b}$ which is the same for all lines. We, therefore, define the base a.c. resistance as

$$
\begin{equation*}
\rho_{11}=\delta \mathrm{V}_{b} \delta \mathrm{I}_{b} \tag{3}
\end{equation*}
$$

for constant $I_{c}$.
The equation for the $I_{c}=0$ line is

$$
\mathbf{I}_{b}=\mathbf{I}_{b}^{\prime}+\mathbf{V}_{b} / \rho_{11}
$$

We now define a reverse current amplification factor as

$$
\begin{equation*}
b=-\delta \mathbf{I}_{b} / \delta \mathbf{I}_{c} \tag{4}
\end{equation*}
$$

for constant $V_{b}$. This is the ratio of a change of base current to the change of collector current responsible for it. The negative sign comes in because it is convenient to have $b$ as a positive number and $\delta \mathrm{I}_{b} / \delta \mathrm{I}_{c}$ is itself negative since an increase of $\mathrm{I}_{c}$ reduces $\mathrm{I}_{b}$. The complete equation is now

$$
\mathrm{I}_{b}=\mathrm{I}_{b}^{\prime}-b \mathrm{I}_{c}+\mathrm{V}_{b} / \rho_{11}
$$

This is conveniently written as

$$
V_{b}=I_{b} \rho_{11}-I_{b}^{\prime} \rho_{11}+b I_{c} \rho_{11}
$$

This has the same interpretation as the one for the collector circuit and so a circuit to which this applies


Fig. 4. Complete d.c. equivalent circuit of on $n-p-n$ transistor. For a $p-n-p$ transistor all voltages and currents are reversed.


Fig. 5. This diagram as a whole is a graph of a resistance in series with a battery. The part in the first quadrant approximates an n-p.n transistor, while the part in the third quadrant approximates a p-n-p transistor.
has the form shown in Fig. 3(b). However, since $b \mathrm{I}_{e} \rho_{12}$ has the opposite sign to $a \mathrm{I}_{b} \rho_{22}$ it is connected the other way round, to oppose the external driving voltage.

The two circuits of Fig. 3 represent the input and output parts of the transistor and together, as in Fig. 4, they form one equivalent circuit of the $n-p-n$ transistor. It is valid for d.c. conditions and accurate in so far as the linear approximations to the transistor characteristics are accurate.

## The p-n-p Transistor

We have now to consider the $p-n-p$ transistor. It is an experimental fact that its characteristics are of the same form as those of the $n-p-n$ but with all the positive signs changed to negative. Because of this, the characteristics are usually drawn upside down compared with Fig. 2. Because all the signs for voltage and currents are reversed, it follows that the same equations apply with the signs of all voltages and currents reversed, and so the equivalent circuit of Fig. 4 also applies to a $p-n-p$ transistor if all voltages and directions of current flow are reversed.

It should be noted that the reversal of signs applies only to the currents and voltages. The quantities $\rho_{11}, \rho_{22} a$ and $b$ still remain positive. Each of these terms is the quotient of a voltage by a current or the ratio of two currents and the quotient or ratio of two negative quantities is positive.

All this may be a little clearer from Fig. 5. Viewed as a whole, the diagram represents a resistance in series with a battery. The line $I_{b}=0$ is for a resistance alone and is merely the graphical representation of a simple resistance. Inserting a battery in series with it to aid the external voltage $\mathrm{V}_{c}$ shifts the line upwards; inserting it to oppose the external voltage shifts it downwards.


Fig. 6. This diagram illustrates the transformation of Fig. 4 into a different equivalent form. At (a) $\mathrm{bl}_{c} \rho_{11}$ is made common to both base and collector circuits, but its effect in the latter is nullified by the extra bottery of the same value opposing it. In (b) this common battery is replaced by a resistance $\rho_{e}$, the voltage being developed across it by the current through it.

The characteristics shown by full lines in the top right-hand quadrant are by themselves those of an ideal $n-p-n$ transistor. Those in the bottom left-hand corner are those of an ideal $p-n-p$ transistor. The dotted parts are non-existent for transistors. The whole diagram can be expressed by one equation which, using the transistor symbols, is

$$
\mathrm{I}_{c}=a \mathrm{I}_{b}+\mathrm{V}_{\mathrm{c}} / \rho_{22}
$$

For the $n-p-n$ transistor $V_{c}$ is positive and $I_{c}$ is positive and only the top right-hand part of the diagram exists. Theoretically, $\mathrm{I}_{b}$ can then be negative as long as $a \mathrm{I}_{b}<\mathrm{V}_{c} / \rho_{22}$, but usually $\mathrm{I}_{b}$ is positive. For a $p-n-p$ transistor, $V_{c}$ and $I_{c}$ are negative and only the bottom left-hand part of the diagram exists. Again, $I_{b}$ can be of opposite sign and so positive, but it is usually negative.

One can see, however, quite clearly that $\rho_{22}$ and $a$ must be positive with both types of transistor for, with parallel straight lines, the values of these quantities are independent of the part of the diagram at which they are taken.

Reverting now to the equivalent circuit of Fig. 4, it is possible to transform this to a different arrangement of its parts which is sometimes convenient. The first step in doing this is shown in Fig. 6(a). The only changes that we have made here are to move the positive terminal of $a \mathrm{I}_{b} \rho_{22}$ from earth to the junction of $I_{b}^{\prime} \rho_{11}$ and $b I_{c} \rho_{11}$ and to insert another battery $b \mathrm{I}_{c} \rho_{12}$ in series with it. These make no difference, for the two batteries $b \mathrm{I}_{c} \rho_{11}$ oppose each other in the collector circuit, so the positive terminal of $a \mathrm{I}_{b} \rho_{22}$ is still at earth potential.

The next step is to replace the $b \mathrm{I}_{c} \rho_{11}$ battery common to both circuits by a resistance and to arrange matters so that the voltage drop produced by the currents $I_{b}$ and $I_{r}$ produces the necessary e.m.f. The circuit then takes the form shown in Fig. 6(b). If (a) and (b) are to be identical, then all similar voltages and currents must always be the same for each. The conditions for identity are then easily found by writing the mesh equations and equating the coefficients of the same currents and voltages.

For Fig. 6(a) (or Fig. 4), we have for the first mesh

$$
\mathrm{V}_{b}=\mathrm{I}_{b} \rho_{11}-\mathrm{I}_{b}^{\prime} \rho_{11}+b \mathrm{I}_{c} \rho_{11}
$$

and for Fig. 6(b) we have

$$
\mathrm{V}_{b}=\mathrm{I}_{b}\left(\rho_{b}+\rho_{c}\right)-\mathrm{I}_{b}^{\prime} \rho_{11}+\mathbf{I}_{c} \rho_{e}
$$

Therefore, $\rho_{11}=\rho_{b}+\rho_{e}$ and $b \rho_{11}=\rho_{e}$.
For the second mesh we have for Fig. 6(a) (or Fig. 4)

$$
\mathrm{V}_{c}=\mathbf{I}_{c} \rho_{22}-\mathbf{I}_{c}^{\prime} \rho_{22}-a \mathbf{I}_{b} \rho_{22}
$$

while for Fig. 6(b) we have

$$
\mathbf{V}_{c}=\mathbf{I}_{c}\left(\rho_{c}+\rho_{e}\right)-\mathrm{I}_{c}^{\prime} \rho_{22}-\mathrm{I}_{b} \rho_{m b}+\mathbf{I}_{b} \rho_{e}
$$

hence $\rho_{22}=\rho_{c}+\rho_{e}$ and $a \rho_{22}=\rho_{m}-\rho_{e}$
Collecting these results, we find that for identity we must have

$$
\begin{aligned}
& \rho_{b}=\rho_{11}-\rho_{e} ; \quad \rho_{e}=b \rho_{11} ; \\
& \rho_{c}=\rho_{22}-\rho_{e} ; \quad \rho_{n b}=a \rho_{22}+\rho_{e}
\end{aligned}
$$

The resistance $\rho_{m}$ may be a little confusing at first since it does not appear in Fig. 6 as a resistance. It is, perhaps, better to regard $\rho_{m}$ as a multiplier for $\mathrm{I}_{b}$ which has the dimensions of a resistance and which when multiplied by $I_{b}$, gives the magnitude of the internal e.m.f. of the transistor.

The circuit of Fig. 6, like that of Fig. 4, represents the static characteristics of an $n-p-n$ transistor in so far as the linear approximation to those characteristics is valid. The circuit for a $p-n-p$ transistor is the same, but with the polarities of all voltages and the directions of all currents reversed.

The usual equivalent circuit is for a.c. conditions only and we could have derived this directly. However, some confusion between the two kinds of junction transistor is avoided if the d..c equivalent circuit is first derived as a stepping stone, and we shall actually find a use later on for the d.c. circuit.

The first step in producing an a.c. equivalent circuit is to produce one which is valid for both d.c. and a.c. We can then get the a.c. one merely by taking away the d.c. part.

Hitherto, the circuits have been, strictly, d.c. ones and we obtain the combined a.c. and d.c. equivalents exactly as we did for the valve by letting each voltage or current be equal to the sum of a mean d.c. component and an a.c. component. Using the sub-
script " $m$ " to denote this mean component and a small letter for the a.c. component, a voltage, say $\mathrm{V}_{b}$, will become $\mathrm{V}_{b m}+v_{b}$. In Fig. 4, therefore, we replace $\mathrm{V}_{b}$ by $\mathrm{V}_{b m}+v_{b}, \mathrm{I}_{b}$ by $\mathrm{I}_{b m}+i_{b}, \mathrm{~V}_{c}$ by $\mathrm{V}_{c m}+$ $v_{c}$ and $\mathrm{I}_{c}$ by $\mathrm{I}_{c m}+i_{c}$. The currents $\mathrm{I}_{b}^{\prime}$ and $\mathrm{I}_{c}^{\prime}{ }_{c}$ are pure d.c. quantities and are not affected.

We can now draw the complete equivalent circuits for Fig. 1 as in Fig. 7 and these are valid for both d.c. and a.c. conditions within the limits imposed by the linear approximation. Notice that in both Figs. 1 and 7 the input voltage $v_{1}$ is of opposite polarity with respect to earth for the $p-n-p$ transistor (b) compared with the $n-p-n$ transistor (a). This is done for simplicity, so that the positive half-cycle of input voltage in both cases acts to assist $\mathrm{E}_{b}$.

Exactly as we did with the valve, we can now drop all the d.c. terms from Fig. 7 to leave only the a.c. ones, and we then get Fig. 8. It is very important to notice that the two circuits (a) and (b) are essentially identical. All the voltage generators of (a) are reversed in polarity in (a) compared with (b), and so are all the directions of the currents. If, therefore, we reverse the polarity of $v_{1}$ in (b) so that it is positive


Fig. 8. The a.c. equivalent circuits of Fig. I are shown here. Since all voltages and currents in (b) are reversed as compared with (a), the two circuits are identical.

Fig. 7. The full a.c. and d.c. equivalent circuits of Fig. i are shown here for (a) on n-p-n transistor and (b) a p-n-p type. The a.c. inputs are of opposite polarity in the two.



Fig. 9. An alternative equivalent circuit to that of Fig. 8 is shown here. This is the one most often used for the transistor.


Fig. 10. A valve circuit which has on equivalent circuit the same as that of Fig. 8 and which, therefore, represents a tronsistor.
to earth as it is in (a), the direction of $i_{b}$ must be reversed and this will reverse the polarity of the generator $a \rho_{22} i_{b}$ and hence the direction of $i_{6}$. In turn, the polarity of the generator $b \rho_{11} i_{e}$ will reverse.

It follows that the a.c. equivalent circuit is exactly the same for both $n-p-n$ and $p-n-p$ transistors.

This equivalent circuit can actually be used to represent, not only junction transistors, but any device whose characteristics can be expressed by two families of curves similar to Fig. 2. It can, therefore, also be used for the point-contact transistor (although it is not always easy to obtain these curves for this device) and for a positive-grid thermionic valve. For a negative-grid valve the grid current is zero, so $\rho_{11}$ is infinite and $b$ is zero; the left-hand half of Fig. 6 then disappears. The right-hand half forms the usual valve equivalent circuit if $\mu v_{g}$ replaces $a i_{b} \rho_{22}$.

There are a great many alternatives to Fig. 8. Some of them are just ordinary circuit transformations of Fig. 8; others are similar but depend on other quantities than $\rho_{11}, \rho_{22}, a$ and $b$ for defining the transistor characteristics. A particularly common alternative to Fig. 8 is shown in Fig. 9; it is the a.c. version of Fig. 6(b). The relations between the two are easily obtained but, as the procedure is exactly the same as for the d.c. circuit and the relations so derived are the same, it is unnecessary to repeat it.

The form of equivalent circuit shown in Fig. 9 is actually the one most used in the literature, although it is usually derived for the earthed-base circuit rather than the earthed-emitter. Generally, the one of Fig. 8 is to be preferred because it leads to simpler design equations. The one of Fig. 9, however, has the merit that it is an obvious equivalent circuit of a thermionic valve plus a pair of resistances. In other words, a real valve circuit can be built which will simulate an earthed-emitter transistor.

In Fig. 9, the elements $\rho_{c}$ and $i_{b} \rho_{m b}$ are the same as the equivalent circuit of a valve if $\rho_{c}=r_{a}$ and $i_{b} \rho_{m}=$ $\mu v_{g}$. If they are replaced by a valve, $\rho_{\theta}$ becomes a cathode resistance and $\rho_{b}$ a grid-cathode resistance
across which $v_{g}$ is developed by $i_{b}$, so $v_{g}=i_{b} \rho_{b}$, and therefore, $\mu=\rho_{m} / \rho_{b}$. The circuit thus has the form shown in Fig. 10.

This circuit is a good one as an analogy for an $n-p-n$ transistor for its characteristics are obvious to anyone versed in valve circuits whereas those of a transistor are not so readily apparent.

It is instructive to insert some numerical values. A junction transistor may have $\rho_{b}=750 \Omega, \rho_{6}=$ $45 \mathrm{k} \Omega, \rho_{e}=35 \Omega$ and $\rho_{m}=1.5 \mathrm{M} \Omega$. The equivalent valve may thus have a resistance of $45 \mathrm{k} \Omega$ which is quite feasible and a $\mu$ of $1,500,000 / 750=2,000$, which is rather impracticable, for it means a mutual conductance of $2,000 / 45=44.5 \mathrm{~mA} / \mathrm{V}$.

The transistor is thus equivalent to a superlatively good valve spoilt by a very low input resistance. It is so much spoilt, in fact, that the overall gain is no better than that of quite an ordinary valve.

So far, we have said little about the point-contact transistor. In fact, the same method of approach is possible and the same equivalent circuits are applicable. In practice, however, difficulties arise in obtaining characteristics of the same form as those of Fig. 2 because, in the earthed-emitter connection, the pointcontact transistor can have negative input and output resistances. It needs careful use in this circuit if it is to be stable.

Because of this, it is usual to plot its characteristics for the earthed-base connection, in which it is inherently stable, and, because this is done for the point-contact transistor, it is quite common to do it also for the junction types. Transistor characteristics and constants are more often published for the earthed-base connection than for the earthed-emitter, however the transistor may be used. It is necessary, therefore, to consider the earthed-base transistor in some detail and this we shall do in Part 3.
(To be continued)

## I.E.E. Iwards to Inthors

THE major premium of the Institution of Electrical Engineers for a paper read or accepted for publication during the last session-the Institution Premium (value $£ 50$ )-is to be given to Dr. D. M. MacKay, of London University, author of "High-speed electronic-analogue computing techniques." The John Hopkinson premium (£25) goes to Dr. M. J. Kelly and G. W. Gilman (Bell Telephone Labs.) and Sir Gordon Radley and R. J. Halsey (G.P.O.), authors of the paper "A transatlantic telephone cable.,"

Dr. N. W. Lewis (G.P.O.) is awarded the Blumlein-Browne-Willans premium ( $£ 20$ ) for his paper "Waveform responses of television links." The Fahie premium (£10) goes to J. M. C. Dukes (S.T.C.) for "The effect of severe amplitude limitation on certain types of random signal," and the Webber premium (£10) to W. E. Willshaw (G.E.C.), Dr. H. R. L. Lamont (R.C.A.) and E. M. Hickin (G.E.C.) for "Experimental equipment and techniques for a study of millimetre-wave propagation."
The premiums to be awarded for papers presented to the Radio Section are: Duddell (£20) to E. G. Rowe, P. Welch and W. W. Wright (S.T.C.) for "Thermionic valves of improved quality for Government and industrial purposes"; Ambrose Fleming (£10) to Dr. P. E. Axon, C. L. S. Gilford and D. E. L. Shorter (B.B.C.) for "Artificial reverberation"; a $£ 10$ premium to H. Page and G. D. Monteath (B.B.C.) for "Vertical radiation patterns of medium-wave broadcasting aerials"; a $£ 5$ premium to M. W. Gough (Marconi's) for "Some features of v.h.f. tropospheric propagation," and another £5 premium jointly to R. C. Glass, G. D. Sims and A. G. Stainsby (G.E.C.) for " Noise in cut-off magnetrons."

# Transmitting Iriformation 

## Colour

SIGNALLING TECIINIQUE IN

THE BRITISH N.T.S.C. SYSTEM

VANY people are of the opinion that when a colour television system is finally adopted for this country it will be a version of the American N.T.S.C. compatible system, scaled down to fit British standards. ${ }^{\star}$ The ultimate choice of system largely depends, of course, on the recommendations of the Television Advisory Committee, and they will no doubt have several alternatives to consider when the time comes. One possible candidate, for example, could be the frame-sequential system and another the version of the N.T.S.C. system in which the colour information is transmitted in a different channel (or even band) from the brightness information. Be that as it may, the T.A.C. have already voiced the opinion (some may think quite wrongly) that British colour television ought to be compatible, which is as good as saying, in the present state of the art, that it ought to be something very much like the N.T.S.C. system, with the colour transmitted either inside or outside of the monochrome band. Undoubtedly this would be popular with the B.B.C., because it would involve very little change to their existing black-and-white transmitting equipmentonly the addition of colour circuits. It may be significant, too, that the two major demonstrations
of colour television so far, by the rival firms Marconi and E.M.I., have both been using compatible systems.

On the assumption that a version of the N.T.S.C. system may well be adopted in Britain (and Wireless World is not necessarily in sympathy with the idea), it would perhaps be worth while looking at some of the aspects of the system which have not been fully explained in this journal so far. These are mostly to do with the processing of the colour information for transmission-a business which is quite significant in that it has some bearing on the design of domestic colour receivers. The transmission of the brightness information is not so important because it is virtually the same as the transmission of black-and-white pictures in our existing system-and in fact the brightness channel of colour receivers would look very much like the whole of the present monochrome receivers.

Wireless World has already explained the basic principles of the N.T.S.C. type of system ${ }^{\star}$ but it may be as well to recapitulate some of the characteristic features. At the transmitting end the colour information from the colour camera is separated into two main components, a signal conveying the brightness information of the picture and a signal

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Fig. I. Block schemotic (very simplified) of the transmitting end of the British N.T.S.C. system. The section to the right of the broken line contains the circuits for producing the modified colour-difference signals from the original ones (to the left of the line)
carrying the colour information (i.e. hue and saturation). The brightness signal is received on existing monochrome television sets as a black-and-white picture, while in colour receivers it is recombined with the colour signal to produce a complete colour picture. We explained how the colour information was produced in the form of two colour-difference signals, $\mathrm{E}_{R}-\mathrm{E}_{Y}$ and $\mathrm{E}_{B}-\mathrm{E}_{Y}$ (as shown on the left of Fig. 1), which were transmitted as amplitude modulation on two components of a sub-carrier having a phase difference of $90^{\circ}$ between them. The two components were combined, as shown by the vector diagram Fig. 2(a), to produce a single colour signal which carried hue information by phase modulation and saturation information by amplitude modulation. (Fig. 2(b) shows the various phase angles of the colour signal and the colours which they represent.) The sub-carrier was actually suppressed and the remaining sidebands were transmitted within the same band as the brightness signal by a process of frequency interleaving.

This description was actually of an early version of the N.T.S.C. system. The system which is now being used is basically the same, but the two colourdifference signals have been slightly modified. They are now each composed of certain proportions of both of the original colour-difference signals. These proportions are arranged so that one modified colourdifference signal conveys colour information ranging from orange to bluish-green (cyan) and the other colour information ranging from yellowish-green to purple. In the British version of the N.T.S.C. system ${ }^{\star}$ the orange-cyan signal is allowed a bandwidth of about $1 \mathrm{Mc} / \mathrm{s}$ while the yellowish-greenpurple signal is restricted to about $400 \mathrm{kc} / \mathrm{s}$.

## Human Colour Vision

This modification to the system is actually a method of processing the colour information which takes advantage of the limitations of the eye's colour vision to secure the best possible transmission conditions within the limitations of the signalling channel. Our November 1953 article on the N.T.S.C. system has already explained that the human eye is insensitive to colour in very fine detail and only perceives it in the form of brightness changes. The N.T.S.C. system therefore does not transmit the very fine detail of the picture through the colour channel, but leaves it to the brightness channel, which has the full $3-\mathrm{Mc} / \mathrm{s}$ bandwidth available to handle it. With very coarse detail the eye is able to perceive the colours properly, but as the coarse detail is represented by a video signal of low informa-tion-content the bandwidth of the colour channel here can be quite small-in fact about $400 \mathrm{kc} / \mathrm{s}$.

In between these two extremes there is a grade of moderately fine picture detail where the eye is only partly effective in its colour perception. It is aware of the existence of colours but cannot distinguish between them properly-in fact it is partly colour-blind. For example, blue and yellow are confused with grey, brown is difficult to distinguish from crimson, and blue is confused with green. Reddish colours, however, remain clearly distinct from blue-greenish colours, and, in fact, the eye tends to interpret all other colours in terms of these two opposites, or mixtures of them. With this moder-

[^8]ately fine grade of picture detail, then, colour vision degenerates from being a three-colour process to a two-colour process.
The effect can be illustrated in graphical form on the well-known Maxwell colour triangle, Fig. 3. Here the three primary colours are at the three corners, while other colours produced from mixtures of them can be specified by the spatial positions of points within the triangle. With two-colour vision, this diagram becomes nothing more than a straight line, as shown, running from orange to cyan across the original triangle. All possible colours in this two-colour system can then be specified by distances along the line instead of by spatial positions within the two-dimensional diagram. (It is interesting to note that orange and cyan are the two basic colours which have been found by experience to give the best approximation to reality in two-colour photography.)

The orange-cyan line, then, represents the " working characteristic" of the eye on colour detail of medium fineness. In the British N.T.S.C. system this particular grade of detail is represented by a band of video frequencies of up to about $1 \mathrm{Mc} / \mathrm{s}$. The transmission circuitry is therefore arranged so that colours in the orange-cyan range are conveyed by a particular signal-the modified colour-difference signal, as mentioned above-which is, in fact, allowed this $1-\mathrm{Mc} / \mathrm{s}$ bandwidth. The other modified colourdifference signal conveys the remaining range of colours, yellowish-green to purple (. 1so shown in Fig. 3) which, when combined with the orangecyan range, give complete three-colour reproduction. But as the eye is only sensitive to three-colour reproduction in very coarse detail it is possible to transmit this yellowish-green to purple signal with quite a narrow bandwidth, in fact about $400 \mathrm{kc} / \mathrm{s}$. Thus, considering the system as a whole, one can see that the transmitted picture information is divided into three categories-coarse picture detail (up to $400 \mathrm{kc} / \mathrm{s}$ ) which is transmitted in full colour; moderately fine detail (up to $1 \mathrm{Mc} / \mathrm{s}$ ), transmitted in two colours; and very fine detail (up to $3 \mathrm{Mc} / \mathrm{s}$ ) which is only transmitted as brightness changes as in an ordinary monochrome system.

The actual method of producing the modified colour-difference signals can be seen from Fig. 1. In the first place, the brightness signal, $\mathrm{E}_{Y}$, which provides the black-and-white picture for existing receivers, is formed by adding together certain proportions of all three primary-colour-component signals from the camera. The colour difference signals $\mathrm{E}_{R}-\mathrm{E}_{Y}$ and $\mathrm{E}_{B}-\mathrm{E}_{Y}$ are obtained, as explained in the November 1953 article, by adding a phaseinverted version of $\mathrm{E}_{Y}$ (that is $-\mathrm{E}_{Y}$ ) to $\mathrm{E}_{R}$ and $\mathrm{E}_{B}$, the purpose being to remove the redundant brightness information which exists in $\mathrm{E}_{R}$ and $\mathrm{E}_{R}$. Green is not transmitted separately because it can be obtained at the receiver by subtracting the sum of the red and blue signals from the brightness (or "white ") signal.
More adding circuits are now brought into play to form the modified colour-difference signals from the original ones. First of all the orange-cyan signal. To produce this, $74 \%$ of $\mathrm{E}_{R}-\mathrm{E}_{Y}$ is added to $27 \%$ of a negative quantity, $-\left(\mathrm{E}_{B}-\mathrm{E}_{Y}\right)$, which is obtained by a phase inverter from the positive $\mathrm{E}_{B}-\mathrm{E}_{Y}$ signal. This - $\left(\mathrm{E}_{B}-\mathrm{E}_{Y}\right)$ signal in fact represents the opposite or complementary colour of blue, which is yellow, and one can see that $27 \%$ of yellow added to $74 \%$ of red

from $E_{R}-E_{Y}$ because it is a resultant of $E_{R}-E_{Y}$ and $E_{B}-E_{Y}$ and although $E_{\mathrm{B}}$ is zero the $-E_{\mathrm{Y}}$ makes $\mathrm{E}_{\mathrm{B}}-E_{\mathrm{Y}}$ into a minus quantity.)
will produce orange. Thus the modified colourdifference signal coming out of the adding circuit now represents the orange in the picture (or colours seen as orange) directly, whereas before it was conveyed indirectly by two separate components in the $\mathrm{E}_{R_{1}}-\mathrm{E}_{Y}$ and $\mathrm{E}_{B}-\mathrm{E}_{Y}$ primary colour channels.
${ }_{R}$ It is easy to see that yellow and red added together will produce orange, but what about the cyan end of the "working characteristic" in Fig. 3? When colours in this part of the spectrum (or colours which the eye interprets as cyan) appear in the picture $\mathrm{E}_{R}-\mathrm{E}_{Y}$ becomes a negative quantity - $\left(\mathrm{E}_{R}-\mathrm{E}_{Y}\right)$, which represents the opposite of red, or peacock blue, while the $-\left(\mathrm{E}_{B}-\mathrm{E}_{Y}\right)$ goes back to positive, which of course represents blue. The added proportions of the two then produce cyan, which lies between blue and peacock blue on the colour triangle. In other words, the colour signal vector which represented orange in the first instance, at " 11 o'clock" in Fig. 2(b), undergoes a $180^{\circ}$ change of phase and appears at " 5 o'clock ", $^{\prime}$ between blue and peacock blue. For the other modified colour-difference signal (yellowish-green to purple) the two constituents are $41 \%$ of $\mathrm{E}_{B}-\mathrm{E}_{y}$ plus $48 \%$ of $\mathrm{E}_{R}-\mathrm{E}_{\gamma}$. When these values are positive they add to give the purple (" 2 o'clock" on Fig. 2 (b)) and when they are negative they give the yellowish-green (" 8 o'clock" on Fig. 2 (b) ).

After the two signals are formed they are passed through band-pass filters as shown. The result is that both of them are effective and give three-colour reproduction for video frequencies up to $400 \mathrm{kc} / \mathrm{s}$ (coarse picture detail), while at video frequencies between $400 \mathrm{kc} / \mathrm{s}$ and $1 \mathrm{Mc} / \mathrm{s}$ (moderately fine detail) only the orange-cyan signal is transmitted, giving the two-colour reproduction to which the eye is physiologically restricted in this range. As was mentioned above, the two colour-difference signals in the earlier N.T.S.C. system were transmitted by being modulated on to two components of a subcarrier, displaced $90^{\circ}$ in phase, which were then combined to form a single r.f. signal (Fig. 2 (a) ). The modified colour-difference signals in the British system are handled in exactly the same way, the orangecyan signal being designated $\mathrm{E}_{I}$ and the yellowishgreen to purple signal $\mathrm{E}_{Q}$.

At this point one might be inclined to ask: why bother to restrict the $\mathrm{E}_{Q}$ signal to about $400 \mathrm{kc} / \mathrm{s}$ bandwidth when the $\mathrm{E}_{I}$ signal is already causing the
final r.f. colour signal to take up a full $1 \mathrm{Mc} / \mathrm{s}$ ? The answer to this is bound up with the problem of avoiding mutual interference and cross-talk between the various component signals in the complete transmission channel (see Fig. 4).
First of all it is essential that the colour signal, which is transmitted within the same $3-\mathrm{Mc} / \mathrm{s}$ band as the brightness signal, shall have minimum visibility on the screens of black-and-white receivers. One of the expedients necessary to achieve this is to make the colour sub-carrier frequency as high as possible so that the pattern it produces on the screen will have a very fine structure. Now with a high subcarrier frequency, placed well to the right as in Fig. 4, it is clearly only possible to transmit the lower sidebands of a $1-\mathrm{Mc} / \mathrm{s}$ colour signal in full, the upper sidebands being partly removed by the upper limit of the $3-\mathrm{Mc} / \mathrm{s}$ pass band. If both $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$ were given the full $1-\mathrm{Mc} / \mathrm{s}$ bandwidth they would both have sidebands like $\mathrm{E}_{I}$ in Fig. 4. Transmission of these two signals on the same sub-carrier would be satisfactory over the small band where double-sideband operation is possible, but beyond this point the missing upper sidebands would have the effect of introducing spurious signals into the $\mathrm{E}_{I}$ channel from the $\mathrm{E}_{Q}$ channel and vice versa. The result on the receiver screen would then be incorrect colour reproduction at the edges of objects.

## Overcoming Colour Cross-Talk

By restricting the $\mathrm{E}_{Q}$ signal to a bandwidth of about $400 \mathrm{kc} / \mathrm{s}$, however, this cross-talk problem is overcome, simply because over the range where $\mathrm{E}_{I}$ consists only of single (lower) sidebands ( $400 \mathrm{kc} / \mathrm{s}$ to $1 \mathrm{Mc} / \mathrm{s}$ ), there is no $\mathrm{E}_{Q}$ signal for it to interfere with. In other words, the two colour-difference signals $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$ are transmitted together on the same sub-carrier only in the video frequency range where double-sideband operation is possible for both, and consequently no interference occurs between them. The sub-carrier frequency is positioned so that this d.s.b. range is big enough to accommodate the band of video frequencies for which three-colour reproduction is effective. Beyond $400 \mathrm{kc} / \mathrm{s}$ a single colour-difference signal is transmitted, $\mathbf{E}_{I}$, in the form of a set of lower sidebands, and only this one signal is necessary because the viewer's eye cannot perceive anything more than the two-colour information which it conveys. In the British N.T.S.C. system the sub-carrier is placed at $2.66 \mathrm{Mc} / \mathrm{s}$, and


Fig. 4. Frequency characteristic of the 3-Mc/s vision channel of the British N.T.S.C. system, showing how the sub-carrier and its sidebands carrying the colour information are placed relative to the upper limit of the pass band.


Fig. 5. Vector diagram of the $E_{I}$ and $E_{Q}$ signals and their resultant colour signal, showing how the colour signal vector can also be separated into components equivalent to $E_{\mathrm{R}}-E_{\mathrm{Y}}$ and $E_{\mathrm{B}}-E_{\mathrm{Y}}$. The angles taken up by the colour signal vector with the various colours are the same as in Fig. 2 (b).
this represents a compromise between two conflicting requirements-mitigating the interference pattern on the screens of black-and-white receivers and avoiding colour cross-talk.

Colour information in the British N.T.S.C. system, then (as in the American system), is transmitted not in terms of the well-known primary colours red, green and blue, but by another group of colours which are more suited to the characteristics of the signalling channel (including the human eye). In the earlier system the red, green and blue primarycolour information was converted into two colourdifference signals which were finally combined into a complete colour signal; this represented the hue of the colour by its phase angle and the saturation by its amplitude. The modified primaries in the later system are also transmitted as colour-difference signals in a similar fashion and the transmitted colour signal has the same phase angle and amplitude as before for a given hue and saturation-that is, for video frequencies up to about $400 \mathrm{kc} / \mathrm{s}$. Between $400 \mathrm{kc} / \mathrm{s}$ and $1 \mathrm{Mc} / \mathrm{s}$ only one of the modified colour-difference signals is in operation, giving two-colour reproduction along the orange-cyan line in Figs. 2(b) and 3. The vector representing the transmitted colour signal then does not rotate through all the positions shown in Fig. 2(b), but rises and falls along the line marked ORANGE-CYAN (in both positive and negative directions).

As far as the receiver is concerned, then, it is presented with a colour signal which is modulated in both phase and amplitude, and from these modulations the original colour information has to be recovered. The technique usually adopted makes use of synchronous detectors. The incoming sidebands are mixed with two local oscillations which have the same frequency as the colour sub-carrier but are displaced $90^{\circ}$ in phase (as at the transmitting end).

One mixer then recovers the $\mathrm{E}_{I}$ signal as a product while the other recovers the $\mathrm{E}_{Q}$ signal, and from these the original $\mathrm{E}_{R}, \mathrm{E}_{B}$ and $\mathrm{E}_{G}$ primary-colour-component signals are eventually obtained.

It is important, however, that the two local oscillations shall have the same phases as the two sub-carrier components at the transmitting end, otherwise they will not recover the $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$ signals but something else. The point is illustrated by the vector diagram Fig. 5, which shows the final colour signal and the $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$ signals from which it is composed. Here, as in Fig. 2(a), the colour signal vector is a resultant formed by the two component vectors $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$, and these component vectors indicate the phases which the oscillators at the receiving end must have if $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$ are to be recovered. It is clear, however, that the resultant vector can be separated into many other pairs of components, apart from $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$, all having the $90^{\circ}$ phase displacement between them. In practice this can be done at the receiver by altering the phases of the two synchronous-detector oscillators -and, in fact, these phases can be shifted so that the detectors recover not $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$ but two components equivalent to $\mathrm{E}_{R}-\mathrm{E}_{Y}$ and $\mathrm{E}_{B}-\mathrm{E}_{Y}$ in Fig. 1, the original colour-difference signals.
Thus, at the receiving end, the colour information can be obtained in two different forms, either as $E$, and $\mathrm{E}_{Q}$ signals giving wide-band colour information, or as $\mathrm{E}_{R}-\mathrm{E}_{Y}$ and $\mathrm{E}_{B}-\mathrm{E}_{Y}$ signals giving narrowband colour information. The advantage to be gained by recovering the narrow-band signals is simplicity of receiver design and hence cheapness in manufacture, and most of the present American receivers are, in fact, using this system. Moreover, R.C.A. in America have recently produced an improved colour demodulator for their receivers which not only recovers the $\mathrm{E}_{R}-\mathrm{E}_{Y}$ and $\mathrm{E}_{B}-\mathrm{E}_{Y}$ signals but the $\mathrm{E}_{G}-\mathrm{E}_{Y}$ signal as well, and also gives enough output to drive the tri-colour c.r. tube directly. It uses only one double valve and is undoubtedly a great advance in circuit simplification. At the same time, of course, this type of operation will not give the high-definition colour reproduction that can be obtained by using more elaborate circuitry to recover the $\mathrm{E}_{I}$ and $\mathrm{E}_{Q}$ signals. However, if the colour information is transmitted in the way described above it will be possible to use the cheaper receivers (a real necessity for this country) and at the same time leave the door open for more complex receivers giving better colour reproduction if they are ever wanted.

## Broadeast Iteceiver Sales

AMENDED figures for the sale of broadcast receivers in April (see p. 346, July) have now been issued by the British Radio Equipment Manufacturers' Association, together with the figures for May. It will be seen from the table that the retail sales of television receivers in May dropped by $11,000(15 \%)$ compared with the previous month and by nearly $40 \%$ compared with the January figure.

|  |  |  | Sound | Radiograms | Television |
| :--- | :--- | :--- | :---: | :---: | :---: |
| January | $\ldots$ | $\ldots$ | 98,000 |  | 35,000 |
| February | $\ldots$ | $\ldots$ | 99,000 | 103,000 |  |
| March | $\ldots$ | $\ldots$ | 93,000 | 23,000 | 98,000 |
| April | $\ldots$ | $\ldots$ | 79,000 | 24,000 | 85,000 |
| May | $\ldots$ | $\ldots$ | 73,000 | 16,000 | 75,000 |
|  |  |  | 15,000 | 64,000 |  |

-And Its Advantages for Band-III Reception

THE present seems an appropriate time to say something about the cascode, because although it is not at all new it has only just begun to be sold to the public. In fact, the said public, as a class, are still blissfully unaware of the infiltration of cascodes into the privacy of the home. All they know is that they are foresightedly taking steps to re-establish the ascendency over the Jones's that they lost when the Jones's too installed television. These steps consist either in buying a new model fitted for Band III, or a box of tricks to adapt their present model; and they hope that, by the time the Jones's realize that advertisement TV has actually begun, the waiting list for equipment to enable it to be seen will be very long. But foresighted though they may be, unless they are technically minded they still won't realize that they have bcugtt a cascode.
Wireless World readers, however, may be more interested in the why and how of the cascode than in knowing which washing powder makes that dainty wooliy newer than new.

The name " cascode" dates from before the war, ${ }^{\star}$ and may perhaps be regarded as an abbreviated form of "casc(aded-triode amplifier having characteristics similar to, but less noisy than, a single pentjode." But that original arrangement was slightly different from the present form $\dagger$ and was devised for an altogether different purpose, at the extreme opposite end of the frequency scale-zero and thereabouts. The cascode has also been highly recommended for audio frequencies. $\dagger$ So altogether it is a versatile creature, and we ought to know something about it.

Fig. I shows a cascode arranged as a resistancecoupled low-frequency voltage amplifier. Essentially the same circuit could be used as the amplifier in a voltage-stabilized power unit. Fig. 2 shows a cascode arranged as a v.h.f. amplifier in a Band-III television receiver or adapter. There are all sorts of variations on the same theme, so the first thing we want to settle is the basic theme itself-the minimum that constitutes a cascode.

The short answer is: two stages of amplification, comprising an earthed-cathode triode, cascade-connected to an earthed-grid triode. But perhaps that answer itself needs a little amplification.

First of all, "cascade-connected." That means connected one after the other, as distinct from valves in push-pull and parallel stages, which come two at a time. It is not-in spite of the appearance of Fig. 1 -quite the same thing as series connection. In cascade connection, which is the usual way of connecting successive valves in an amplifier, the output of the first provides the input of the second.

Now there are three basic ways of connecting a

[^9]single valve to make a stage of amplification. These ways can conveniently be named according to the electrode that is earthed; the other two electrodes are those used for input and output respectively. The commonest arrangement of the three is the earthed-cathode, in which the input is led to the grid and the output taken from the anode. The next commonest is the earthed-anode, better known as the cathode follower. It, too, has the input connected to the grid, but the output from the cathode. Lastly, the earthed-grid (called by crypto-Americans


Above: Fig. I. Circuit diagram of a cascode arranged as a resistancecoupled amplifier.


Above right: Fig. 2. Circuit diagram of a cascode arranged as a v.h.f. amplifier, say for Band |II television.


Fig. 3. Essentials of the three ways of connecting a triode as a stage of amplification: (a) earthed cathode; (b) earthed anode; (c) earthed grid.
"grounded-grid"), in which the grid (however did you guess?) is earthed, the input goes to the cathode, and the output comes from the anode. Fig. 3 lines up these three, stripped of all practical details, for inspection.
In a two-valve cascade-connected amplifier, the first stage can be connected in any of these three ways. Each of these three varieties can be subdivided into three, according to the way the second stage is connected. So altogether there are nine possible combinations. When this fact first dawned on my consciousness, I supposed that most of the combinations would be of purely academic interest, only two of three being used in practice; but on going steadily through the lot I found that most, if not all, had been used at some time, for some purpose. For instance, did you know that there was actually a special type of valve-the 6B5-embodying the earthed-anode earthed-cathode combination? Then the earthedanode earthed-grid combination is the basis of the well-known coupled-cathode or " long-tailed pair," used in many valve voltmeters and oscilloscopes. The earthed-grid earthed-cathode is used in radar and other v.h.f. receivers. But however interesting it might be to explore all nine, the only one on to-day's schedule is, as already mentioned, the earthed-cathode earthed-grid combination.

## Coupling Impedance

You will of course want to know why anybody should choose this in preference to the earthed-cathode earthed-cathode, which gives easily the most amplification of any of the nine. But before going into that, let us just make sure that Figs. 1 and 2 really are earthed-cathode earthed-grid cascade combinations and not just the valves connected in series that they appear to be. Looking again at Fig. 3 we see that all three circuits include a resistor (or some other kind of impedance) which performs two roles: providing a path for the steady feed current through the valve, and by its impedance causing the signal current to set up a signal output voltage. This output voltage can then be applied to the input of a following stage. If this following stage is of either the (a) or (b) types, the coupling impedance is still required for the first stage; but type (c) is unique, because not only do its input terminals provide an impedance across which the input voltage can be set up, but the cathode terminal in particular also provides a source of feed current for the first stage. So the coupling impedance can be omitted.

As drawn in Fig. 1, the second or upper valve appears most clearly in its role of coupling impedance and steady-current feeder, corresponding to the resistor in Fig. 3(a). To emphasize its other role of second stage in a two-valve amplifier, I have redrawn Fig. 1 as in Fig. 4. It is now clearly the same as in Fig. 3(c), the only apparent difference being the point to which the grid is connected; and that is the same in principle, being equivalent to earth even though, in order to bias the grid suitably relative to its cathode, it is taken to a source of positive potential. Fig. 4, although its circuit is identical with Fig. 1's, shows more clearly that, while the second valve is in series with the first for current-feed purposes, it is truly in cascade as regards signal amplification.

And now we come to the question of why this arrangement should ever be preferred to the more highly amplifying all-earthed-cathode system. The
answer depends on which of the two main applications we have in mind. The chances just now are heavily in favour of Band III being in mind, rather than voltage stabilizers, though we'll come to them too in due course.

At the very high frequencies of Band III it is difficult to generate such massive power for transmitting as on the lower frequencies. The waves, when transmitted, are more rapidly attenuated; and the receiving aerials are necessarily short and therefore limited in collecting ability. So the signal voltage, or rather microvoltage, that can be brought to the input of a receiver is likely to be very small. At the same time it must, for television, be spread over a wide frequency band and so has to compete with a lot of noise of the kind self-generated in circuits and valves. In fact, even though v.h.f. isn't too easy to amplify, the limit is not set by that difficulty so much as by the signal-to-noise ratio that can be achieved. If the signal doesn't succeed at the outset in poking its head high enough above this noise, no amount of subsequent amplification will help it, because the noise is amplified too. So all depends on the first stage of amplification, in which the signal is at its weakest.

Because of the greater difficulties of amplifying v.h.f. signals, there is an obvious inducement to change to a lower frequency as soon as possible. But a frequency-changer stage is at best a comparatively noisy affair, so doesn't make at all a suitable first stage for signals that are already barely strong enough to stand clear of noise. A preliminary stage of amplification, even if it doesn't give very much, and even if it introduces some noise itself, helps the signal to master the greater frequency-changer noise. In case this point is not entirely obvious, let us suppose that a certain frequency changer introduces $6 \mu \mathrm{~V}$ of noise and neither amplification nor loss, and that a stage of amplification introduces five-fold voltage gain and the equivalent of $3 \mu \mathrm{~V}$ of noise at its input. If the incoming signal is $16 \mu \mathrm{~V}$, together with $2 \mu \mathrm{~V}$ noise, the incoming signal/noise voltage ratio is $8: 1$ and its power ratio $64: 1$. After the frequency changer alone the signal would still be $16 \mu \mathrm{~V}$ and the noise voltage $\sqrt{ }\left(2^{2}+6^{2}\right)=6.3^{\star}$, so the signal/noise power ratio would be degraded to 6.5: 1. But if the amplifier were used first, its output would include $5 \times 16=80 \mu \mathrm{~V}$ of signal and $5 \sqrt{ }\left(2^{2}+\right.$ $3^{2}$ ) $=18 \mu \mathrm{~V}$ of noise (nearly $20: 1$ power ratio), and the output of the frequency changer following would be $80 \mu \mathrm{~V}$ signal and $\sqrt{ }\left(18^{2}+6^{2}\right)=19 \mu \mathrm{~V}$ noise, making the ratio $18: 1$. This simplified example shows how even a moderate amplification, itself not noiseless, can much improve the net result.

## What Valve to Use

Having decided to use at least one stage of amplification at the original frequency, we are then faced with conflicting claims in the choice of amplifier. A r.f. pentode has internal screening which helps to prevent feedback via stray capacitance turning the amplifier into an oscillator, as would inevitably happen with an earthed-cathode triode owing to its anode-togrid capacitance. On the other hand, a pentode is a good deal noisier than a triode. I went into detail about this noise business in the May and June 1952 issues, so for present purposes it should be enough

[^10]

Fig. 4. Rearrangement of Fig. I to show more clearly the role of the second valve as an earthed-grid stage.
to say that because the current through a valve consists of separate electrons there is bound to be a small random fluctuation around even the steadiest average, and it is this fluctuation that is referred to as noise. In a pentode the fluctuation is greater than in a triode, because on top of this chance of slightly fewer or more electrons arriving during any instant, there is the chance of whether any particular electron will be intercepted by the second grid or go right on to the anode. Though you might not think so, this extra chanciness adds very considerably to the noise. The average for 10 different kinds of pentodes used for amplification showed $75 \%$ greater noise voltage than the average for a number of triodes. So there is a strong inducement to use a triode, if it can be made to amplify at v.h.f.

As we know, it cannot if it is simply connected in the usual earthed-cathode manner. This is where the earthed-grid triode came in, especially during the last world war for $200-\mathrm{Mc} / \mathrm{s}$ radar. Unfortunately an earthed-grid triode presents a very low input impediance, even by v.h.f. standards. As we see from Fig. 3(c), it is not much more than the output impedance of a cathode follower, and we know that that is very low-approximately $1 / \mathrm{g}_{m}$, or say $200 \Omega$ if $\mathrm{g}_{m}$, the "slope" of the valve, is $5 \mathrm{~mA} / \mathrm{V}$. The only difference between this circuit and a cathode follower is the external anode resistance, which reduces the overall slope by adding to the internal anode resistance, $\mathrm{r}_{a}$. In a v.h.f. amplifier, the external resistance is the dynamic resistance of a tuned circuit, and is likely to be considerably smaller than $\mathrm{r}_{a}$. So the input resistance is still very low, and establishing a signal voltage across it costs plenty of signal power. Stage gain is therefore even less than usual for a triode.

The attraction of the cascode is that it combines the low noise of a triode with earthed-cathode input resistance and the high gain and stability of a pentode. And although the circuit diagrams here make it look as if two valves were being used in place of one, in practice the two triodes take the form of one double triode costing about the same as one pentode and occupying the same space and valve-holderage.

The second half of the cascode is an earthed-grid stage as just described, and its low input resistance constitutes the anode load or coupling resistance ( $\mathrm{R}_{f}$ ) of the first half. So although this first half is an earthed-cathode stage, it is prevented from bursting into oscillation by the fact that its $\mathrm{R}_{a}$ is too low to give enough amplification. When the $\mathbf{R}_{a}$ of a voltage amplifier is much smaller than the $r_{n}$ of the valve, the voltage amplification (A) is approximately equal
to $\mathbf{R}_{a} \mathrm{~g}_{m}$. But in this case we know that $\mathrm{R}_{a}$ is approximately equal to $1 / g_{m}^{\prime}$, where $g^{\prime}{ }_{m}$ is the slope of the second triode, as modified by its anode load. So the amplification of the first stage is

$$
\mathrm{A} \simeq \mathrm{R}_{1} \mathrm{~g}_{m} \simeq \frac{\mathrm{~g}_{m}}{\mathrm{~g}_{m}^{\prime}}
$$

The slopes of the two valves in a double triode are (or should be) more or less equal. So even if the load resistance of the second were as much as equal to $\mathrm{r}_{i l}, \mathrm{~g}^{\prime}{ }_{m}$ would be half $\mathrm{g}_{m}$, and A consequently no more than 2 . In practice, at v.h.f., it would probably be nearer 1. This doesn't mean that the first stage contributes next to nothing; its real job is to drive the second stage without damping down the input circuit or introducing a lot of noise. It functions, in fact, very like a cathode follower.

The output of the second stage is quite conventional. Fig. 2 shows the whole cascode as arranged for Band-III amplification. Instead of being taken to a tapping on a potential divider, the grid of stage 2 has conventional bias arrangements, but at signal frequency is tied down to earth by C . This modification of Fig. 1 enables automatic gain control to be applied in the customary manner to stage 1. In Fig. 1, a.g.c. bias would be almost completely ineffective in changing the anode current and with it the amplification, because the fixed potential of grid 2 would tend to hold it constant. The only other thing that might perhaps excite curiosity about Fig. 2 is L. Its purpose is to neutralize stray capacitance at the top frequency of Band III, where there would otherwise be a falling off in amplification.

## Mathematical Approach

If, as I hope, you like to work things out mathematically for yourself, I can recommend analysing Fig. 1 by writing down the equations-five of themexpressing overall voltage amplification, A; signal anode current, $i_{a}$; signal voltages of the two anodes, $v_{a 1}$ and $v_{a_{2}}$; and signal voltage between second grid and cathode, $v_{g_{2}}$; in terms of the $\mu$ and $\mathrm{r}_{a}$ of the two valves, the signal voltage $v_{g_{1}}$ applied to the first grid, and $R$. If the valves are not assumed to be identical, the answer should come out to

$$
\begin{equation*}
\mathrm{A}=\frac{\left(\mu_{2}+1\right) \mu_{\mathbf{2}} \mathrm{R}}{\mathrm{R}+\mathrm{r}_{a 2}+\left(\mu_{2}+1\right) \mathrm{r}_{a 1}} \tag{1}
\end{equation*}
$$

If the valves zre identical this simplifies to

$$
\mathrm{A}=\frac{(\mu+1) \mu \mathrm{R}}{\mathrm{R}+(\mu+2) \mathbf{r}_{r}}
$$

while if $\mu$ is large enough for the difference between it and $\mu+2$ to be neglected,

$$
\begin{aligned}
& \mathrm{A} \simeq \frac{\mu \mathrm{R}}{\mathrm{R} / \mu+\mathrm{r}_{a}} \\
& \text { which compares with } \\
& \mathrm{A}=\frac{\mu \mathrm{R}}{\mathrm{R}+\mathrm{r}_{a}}
\end{aligned}
$$

in a conventional single-triode amplifier. These results prove what we have already gathered, that the v.h.f. amplification of the cascode as a whole is not noticeably greater than would be given by one of its valves connected as an ordinary stage if there were no such thing as Miller effect to upset its working. But there is, and the addition of the second triode overcomes it without having to fall back on the noisy pentode.

Looking at things from different viewpoints is usually a help in understanding them; and having
considered the cascode as two successive conventional stages in an amplifier, we may now care to think of it as a single unconventional stage. The difference between pentode and triode characteristics can be expressed in one way by saying that the pentode has a very much larger internal resistance $r_{a}$, which is the same thing as saying that its anode current is very little affected by its anode voltage. Putting a load resistance in series with the anode therefore has hardly any effect on the amount of anode current. This is not so with a triode, however. When its anode current is increased by making the grid less negative, any anode load resistance causes the anode voltage to drop, and this cuts down the increase that would otherwise have taken place.

With this in mind, consider Fig. 1 again. The first (lower) valve has an anode load resistance consisting of the other valve plus $R$. If the grid of this other valve had a fixed bias relative to its cathode, the action would be as just described. But because the grid is tied to a cixed potential, the drop in cathode potential caused by any increase in anode current is equivalent to making the grid less negative, which operates to maintain the increase in anode current. Since the cathode potential cannot alter much without drastically altering the anode current in this way, it tends to stay nearly constant. This is the same


Fig. 5. Essentials of the usual type of voltage stabilizer.


Fig. 6. Modified cascode substituted for $V_{1}$ in Fig. 5. The modification consists of $R^{\prime}$, which raises the amplification of the stage by increasing the anode current and consequently the $\mathrm{gm}_{\mathrm{m}}$ of $\mathrm{V}_{1 \mathrm{a}}$.
thing as the lower valve's anode voltage staying nearly constant, regardless of changes in anode current induced by changes in its grid voltage. So the system as a whole behaves something like a pentode, in two ways: the anode current has been made to depend almost exclusively on the voltage of the first grid regardless of the presence or absence of $R$; and the potential of the electrode next to that grid-the first anode-has been made to stay nearly constant, and so to simulate the screen grid in a pentode, which keeps the amplifier stable.

## Voltage Stabilizer Application

There is no reason, of course, why the cascode cannot be used at lower radio frequencies than v.h.f., or even audio frequencies, but amplification of lower r.f. is not usually limited by valve noise, and a pentode gives more amplification. For a.f. it is an interesting question whether, on balance, the cascode is better than the two valves of a double triode connected in the ordinary way as two earthed-cathode stages. Coming at last to the amplifier used in the usual type of voltage stabilizer, there is much more to be said for the cascode. Most of it has been said very well by V. H. Attree, ${ }^{\star}$ who has devised a modification that looks like establishing the cascode as undisputed king of this situation.

Just to make sure that we are both thinking about the same " usual type of voltage stabilizer," Fig. 5 is an outline diagram, for identification, in which $\mathrm{V}_{1}$ is the amplifier in question. The greater the amplification of this stage, the more effective the stabilizer. So the choice of valve is almost invariably a pentode. To extract a reasonable proportion of the valve's potential amplification, the resistance $R$ must be large. But the voltage across it, being the bias for $\mathrm{V}_{2}$, is normally quite small, which means that the valve is working under conditions (to wit, low $\mathrm{g}_{m}$ ) that throttle most of its amplification. If two stages were used, the output voltage would be of the wrong polarity and the stabilizer would become an unstabilizer. Two stages, that is to say, connected in the usual earthed-cathode manner. But this difficulty does not apply to the cascode, because its second stage, being earthed-grid, does not reverse the polarity. The criticisms that have just been made about the pentode would go for the cascode too, if it were as per Fig. 1. But Mr. Attree has pointed out that when $\mu_{1}$ and $\mu_{2}$ in our equation (1) are large (because a high- $\mu$ double-triode has been chosen), and $\mathbf{R}$ and $\mathrm{r}_{a 2}$ are neglected in comparison with $\mu \mathrm{r}_{a 1}$, the thing reduces to

$$
\mathbf{A} \simeq \frac{\mu_{\mathbf{1}} \mathbf{R}}{\mathbf{r}_{a 1}}=\mathbf{g}_{m 1} \mathbf{R}
$$

So the amplification depends hardly at all on the $g_{m}$ of the second triode and almost entirely on that of the first, and it doesn't matter if the current through $\mathbf{R}$ and the second valve is small, so long as it is large enough for a good big $g_{m}$ in the first. The modification therefore consists of an extra resistor to pass more current through this valve.

The circuit (less all frills) is as Fig. 6. With this improvement, the voltage amplification of the cascode can easily be well over 1,000, which makes the stabilizer incorporating it a very good stabilizer.

[^11]
## ELECTRONIC TELEPRINTING

A NEW electron-image tube that can translate coded signals from teleprinter tape or other sources into clearly defined letters and figures at speeds up to 100,000 words per minute for high-speed photographic recording has recently been developed by R.C.A. In operation the tube simulates the process of typesetting by selecting letters and figures one by one from a "font" and placing them in luminous form on the 5 -in circular tube face, either in lines or in any pattern desired. The "font" is actually a lantern slide external to the tube, bearing a chart of letters and figures. An image of this is projected on to a photo-emissive cathode at one end of the tube which emits a stream of electrons in a corresponding pattern. The electron stream is then accelerated forward in the tube by a potential of 100 V applied to the conductive wall coating and focused by an external coil providing an axial magnetic field.
The selection of letters and figures in the required order is accomplished by a small aperture of 0.04 -in diameter at the neck of the tube which permits only one character at a time to pass through. As the electron stream pattern carrying all the letters and figures moves towards this aperture, a magnetic deflection coil (mounted inside the focus coil) shifts the stream so that only the desired character passes through and travels towards the tube face. Another set of coils then focuses and deflects the character to its proper place on the phosphor screen.
As many as 4,000 characters have been produced clearly in a single pattern on the 5 -in tube face. The size of the letters and figures on the screen can be enlarged, however, by the second set of coils if required.
The deflection coils for selecting and positioning the characters contain windings for both vertical and horizontal deflection, as in television. In the teleprinting application these coils are supplied with suitable steps of current in accordance with the coded information which is "read off" (by photo-electric means) from the teleprinter tape.
When the new tube achieves commercial form its initial applications are likely to be in electronic message transmission and in computing systems. Later it may be developed for electronic typesetting.


COLOUR TELEVISION STUDIO, claimed to be the first one built specifically for colour, and recently put into operation by the National Broadcasting Company of America at Burbank, California. An elaborate lighting system with 2,400 controls permits the pre-setting of lighting for ten scenes and also ten changes of lighting within any one scene. An unusual feature is an "oudience pit" which accommodates the studio audience below floor level to ovoid interference with the cameras.


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# RaNDOM Radiations 

By " DIALLIST"

## " Night Starvation"

EVEN in what passes nowadays for summertime, the mains voltage in some localities is liable to fall somewhat after, say, 8 p.m. or a little later. If the fall is not very great and the television receiver is in tip-top form, the effects of this may be hardly noticeable; but with a bigger fall or a set which has been in use for some time without renewals or replacements this "night starvation" can lead to picture shrinkage, with the typical black borders. When that happens the first thing that I suspect is the h.t. rectifier, particularly if it's of the metal type. Metal rectifiers are sound and reliable components, but they don't last for ever and it pays to renew them when tests show that their output is appreciably down. I'd estimate about two years or perhaps a bit more as their normal useful life in an average family set.

## A Radio-link Problem

THE temporary East Anglian television transmitter at Tacolneston reradiates the London transmissions which it receives direct. This has been leading recently to spots of bother caused by r.f. interference from a Continental sound-broadcasting station. I haven't yet been able to find out what station it is. Under certain conditions the interference is fairly mild, merely causing a certain amount of "fish netting" on the picture. But it can at times be bad enough to blot out the picture and to make the accompanying sound almost unintelligible. To prevent viewers in the Norwich area from snowing dealers under with complaints about the misbehaviour of their perfectly guiltess receivers the B.B.C. has very wisely devised a means of letting them know when it's the transmission that is to blame: vertical white bars sent out every so often indicate to the viewer that interference is affecting the transmission.

## A Monitoring Suggestion

This idea might, I think, be carried a good deal further, fo: TV receivers are often blamed by their users for doing things that they can't help doing. Some of the O.B. cameras, for example, "ring" quite severely, producing pronounced white outlines to the right of dark objects. And this
effect can be made worse than ever if a relay is made over a long landline. An occasional word from the announcer about this might save viewers from worrying about their sets and servicemen from having to make unnecessary journeys. I've a feeling that some, at any rate, of the monitoring should be done with ordinary domestic receivers. Those in charge would then be able to see whether any transmission was likely to cause the sets of Smith, Jones, Brown and Robinson to play up and a word of explanation (and of comfort) could be issued at suitable moments. You know the kind of thing I mean: the sync isn't always able to lock one of the scans-or maybe both of them-properly; a change of camera means a fall in the brightness level, or vice versa.

## Over-Simplified?

Yes; I know that these things wouldn't happen if TV sets were a little more elaborate-and, therefore, a little more expensive. But, domestic receivers being what they are, the plain and inescapable fact is that they do happen. And that brings me to the warning given recently to the radio industry by Harold Bishop, to
whom I offer my humble felicitations on his well-deserved Birthday Honour. What he said in effect was that there are limits beyond which simplification and price reduction cannot reasonably be carried by television receiver manufacturers. And there couldn't be a truer word spoken. Many people feel that these things have been taken too far already. Bringing down prices is fine from one point of view but it's a far from unmixed blessing if it entails, for example, lack of d.c. restoration and d.c. amplification, plus synchronizıng arrangements so poor that the picture won't lock unless the signal is bang up to the mark. Myself, I believe that large numbers of folk would gladly pay a bit more for sets that didn't suffer from these shortcomings and whose pictures remained without flutter when aeroplanes were passing by.

## The Tape-Recorder Cult

THE tape-recorder is, I suppose, the lineal descendant of the Dictaphone and other similar machines designed originally purely for office use. Today, tape recording has become a hobby whose addicts are every bit as enthusiastic as were the fans of the early days of wireless. To the musical it is of course a great joy to be able to make a record (I nearly wrote "recording"!) of any outstanding broadcast and to have it available when wanted. I must have been quite an early user of the


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Dictaphone. Years ago I used to answer a vast number of letters. It was, as you may imagine, the greatest possible convenience to be able to get a few off my chest whenever I felt so minded by dictating my replies into the machine. Once, though, there was a sad mishap. While on a fishing holiday in Devon I recorded a big batch and posted the wax cylinders to my typist. Light-hearted P.O. sorters must have had fun and games with the parcel, for it arrived with all the records in more or less powder form. That, anyhow, couldn't happen with tape.

## Balanced or Unbalanced

WHEN television broadcasting was in its youth, I think I'm right in saying that the great majority of receivers were designed for use with balanced twin aerial feeders. To-day all (or very nearly all) use co-axial feeders. I'm told that this means a small reduction in manufacturing costs; if so, I wonder whether it's worth it. Interference seems to become worse and worse and a good deal can be picked up by a co-axial feeder, even though its metallic sleeve is earthed. But balanced twin, with earthed metallic screen and correct impedance matching, picks up little or none; use it and, if need be, a mains suppressor and you'll get no interference except what is actually picked up by the aerial itself., Some time ago I was using an " H " aerial over looft from the nearest road and neary 60 ft above its surface. Motorcar interference was a nuisance with a coaxial feeder in use; a change to balanced twin, with the necessary alterations to the receiver and careful impedance matching made all the difference in the world.


TELEVISION SCREEN MASK for shielding the tube face when viewing in daylight or with the room lighting on. Made from compressed fibre material with a light-absorbent inside lining, it is fixed to the front of the set by rubber suckers and secured by an elastic strap. Available from Vendoma (TV), Station Buildings, Preston Park, Brighton, in sizes for $12-\mathrm{in}, 14-\mathrm{in}$, $15-\mathrm{in}$ and 17 -in receivers.


## BULGIN

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## UNBIASED

## Peccavi

IN the June issue I mentioned that I was going to try to accelerate the growth of plants in my garden by exposing them to the radiations of an r.f. oscillator. I had been interested in an item of news received from the U.S.A. which said it had been noticed that in the neighbourhood of high-powered television transmitters a veritable jungle of luxuriant undergrowth had sprung up.

I found I was quite mistaken in thinking that there was anything new in this as I have been sent several photographs (one of which is reproduced) and data concerning the growth of r.f.-nurtured plants as far back as 1939. The two hyacinths on the left of the picture received radio treatment ( $50 \mathrm{Mc} / \mathrm{s}$ ) and flowered 15 days earlier than the untreated one shown with them. I can only say I feel ashamed of my own ignorance.

In the June issue I also discussed early talking machines and a reader has tried to put me on the penitent's stool for this by implying that I said that needles and not sapphires were invariably used with disc records. To


## Electronics in the garden.

prove me wrong he has sent me an Edison sapphire-using disc but adds "I see that, with true legal caution, you have left yourself a loophole."

What I certainly did not know was that some disc machines used a screwed rod to propel the sound box across the record. I thought that with discs this was invariably done by the needle running in the groove as in modern machines. This is not so, however, and the makers of the record took pride in announcing that the grooves were too shallow to have the task thrown on them of pushing the sound box along; records are made of sterner stuff nowadays.

## Mummified Music

I WAS very impressed by the demonstrations given in the Royal

Festival Hall in May of the degree of hi-fi which can be achieved in mummified music by modern methods of recording and reproduction. It is given to a very few to be able to listen to a chronologically and topographically side-by-side comparison of the real thing and an embalmed version of it. Despite the fact that the hall was packed to capacity, the audience was very small compared with the number who would have liked to be there.

It is unthinkable that there should not be a repeat performance in a few months' time and I have been trying to think of a method whereby a larger audience could be reached. Mrs. Free Grid made the tom-fool suggestion that the B.B.C. should be asked to put it on the air. Apparently it did not occur to her that the musicmangling properties of the ordinary domestic wireless receiver would reduce the living and the mummified performances to the same low level; in fact, the embalmed version could be made to sound better than the real thing, for special recordings could be made to compensate to some extent for the deficiencies of the average set.

In a few years' time when there will be a hi-fi receiver in every listener's home it will be a different thing. But even to-day the number of f.m. set owners is certainly much greater than the seating capacity of the Royal Festival Hall. To hear the programme on v.h.f. would certainly not be the same as hearing it in the flesh.

All the same, I for one am willing to agree not to apply for a ticket for the next demonstration but to listen at home and let my seat be occupied by somebody less fortunate than myself who is at present beyond the reach of v.h.f. either for geographical or financial reasons. I, therefore, appeal to all other f.m. people to make a similar offer and to the organizers of the demonstration to approach the B.B.C. in the matter.

## Electronic Morphimeter

I WAS interested to learn that an electronic device has been developed for gauging the depth of sleep or unconsciousness by measuring the skin resistance, which apparently varies in step with it. Surely this should have many applications other than the medical one mentioned. Apart from its obvious use in the boxing ring where it would enable the referee to see at a glance the exact condition of a recumbent pugilist it supersedes the existing form of baby alarm first described in this journal nigh on thirty years ago.

This old type alarm, as you may


Further outlook unsettled.
remember, consisted of a mike suspended over the cot so that the nasty noises emitted by a baby are conveyed to the doting parents below. With this new device the changing resistance of the baby's skin as it approaches bawling point could obviously be caused to operate the alarm, thus obviating the mental trauma which a well-known psychiatrist has stated that babies receive when their immediate wants are not anticipated.

I would like to point out, however, that this idea of a morphimeter is not quite as new as it seems. The varying skin resistance on which its action depends is caused, I understand, by the fact that the rate of metabolism in the body changes according to the degree of unconsciousness. This changing rate of metabolism not only varies the skin resistance but has the same effect on the dielectric constant of the body.

Long years ago I fixed up a capacitor, the two electrodes of which were the bedspring and the wire resh of an electric blanket which Mrs. Free Grid always uses. Her body was, of course, the dielectric and I found that the capacitance of my crude device varied according to the depth of her sleep. I put this to practical use by causing it to operate an indicating meter downstairs so that on those occasions when I was detained late at the office I could avoid creeping upstairs to bed until I was assured that she was really sunk in a deep sleep.


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he AVO Valve Characteristic Meter Mark III offers the Radio Engineer far more than is generally implied by the words "a valve tester".
This compact and most comprehensive Meter sets a new high standard for instruments of its type. It will quickly test any standard receiving or small transmitting valve on any of its normal characteristics under conditions corresponding to a wide range of D.C. electrode voltages.
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Flate current- 1


Fig. 2
Screen current - plate voltage

Series of pentode choracteristic curves with grid voltage changing 2 volts/step from $+16 \vee$ to below zero, illustrating Type 570 operation with eight positive-bias curves per family. Vacuum tube is a 6AQ5, under these conditions: Plate load... 300 ohms, peak plate voltage... 100 v , screen-grid voltage ... 100 v , vertical scale ... $10 \mathrm{mo} /$ division, horizontal scale, fig. 1,2 , and $3 \ldots 10 \mathrm{v} /$ division, fig. 4, 5, and 6.,. $2 \mathrm{v} /$ division.

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Fig. 6
Grid current-grid voltoge


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#### Abstract

The W97 capacitor, although of diminutive size, is an extraordinarily robust unit. Most miniature units are prone to weakness in end connections and generai mechanical flimsiness. Such undesirable features are eliminated in the W97 by the special processes used and extreme care in manufacture. CAPACITQR UNIT A single metallised paper is used to wind this unit which is made possible by the use of Hunt's Patent covering the "castellated" pattern. Recent development by Hunt on a special impregnating material gives the unit remarkable brackets of operating temperature.

\section*{CASING}

Hunt's patented double metal tube, sealed with the special "Thermetic" compound, provides positive closure on the casing and lead entry, ensuring positive hermetic sealing. INSULATION OF CASING The capacitors are supplied without an insulating medium on the case. If specially requested they can be supplied with,an approved plastic sleeve which increases the dimensions by $0.07^{\prime \prime}$ in length and $0.03^{\prime \prime}$ in diameter. TERMINATIONS The terminations are of 24 gauge tinned phosphor bronze wire having a nominal length of $1 \frac{1}{2}$. Special attention is paid to the retinning of the wires after the capacitor is fully processed. Connection is made to the unit by applying copper spray to the metallising. The pigtail is soldered to this bond giving a perfect connection of exceptional strength.

\section*{INDUCTANCE}

W97 "Thermetic" Midgets have a very high self resonant frequenty-the following figures are quoted as a guide. 50 pF at 600 volts, which is the lowest capacitance in the range, has a self resonant frequency of 280 mega-  maximum capacitance, it is 8.5 megacycles.

\section*{INSULATION RESISTANCE}

This is measured at working voltage at a temperature of $20^{\circ} \mathrm{C}$. The minimum capacitance in the range, 50 pF at 600 yolts, has an insulation resistance greater than $2,000,000$ megohms. The maximum capacitance in the range $0.04 \mu \mathrm{~F}$ at 200 volts, has an insulation resistance greater than 25,000 megohms. The intermediate capacitances are approximately pro rata. POWER FACTOR Less than $2 \%$ at 1,000 cycles per second at 20 C . CAPACITANCE TOLERANCE


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|  |  |  |  |  | $\frac{4}{0}$ | $\begin{aligned} & \text { U } \\ & \text { U } \\ & \text { O } \\ & \text { E } \end{aligned}$ |  |  |  |  |  |
| BR. 1102 | 50 | Th. | 483 | 241 | $8 \cdot 2$ | 230 | 12.0 | $20 \cdot 0$ | 45 | 42 | $20 \cdot 0$ |
| BW. 1102 | 50 | Th. | 473 | 152 | $8 \cdot 2$ | 230 | 12.0 | 20.0 | 45 | 42 | $20 \cdot 0$ |
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GOODMANS CORNER CABINETS for the AXIOM 150 Mark 2 manufactured by us to Messrs. Goodmans' specification and approved by Messrs. Goodmans Height 44in. Price: complete kit in plain board with lin, thick felt, 8 gns. Price: ready buile, 10 gns . Finished in figured walnut 16 gns Other veneers to order. Carriage extra according to area, Quotation by return.

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 FEEDER UNITSNo. I "SYMPHONY" TUNER A T.R.F. model designed for the quality reception of local stations. Quality is adequate for amplifiers of the highest fidelity class. Infinite impedance detection. Controls: gain, wave-change and radio/gram switch. Alluminated engraved glass dial. Latest miniature valves. Overall dimensions Yin, wide $\times 6$ in. deep $\times 6$ in. high. and $250 / 300 \mathrm{v}$. at. $15 \mathrm{~m} / \mathrm{a}$. Price c7/7/-. Carr. and pkg. 5/.
No. 2 "SYMPHONY" SUPERHET TUNER. Three wave bands, advanced circult, very new est valve types, floodlit glass dial with bronze escutcheon provided Suitable for use with the best amplifiers. Overall dimensions
 deep. Controls: on/off/gain, radio/ gram, wave-change
Dial cut-out: 8 and $x$
$4 \frac{1}{2}$ in. reading horizontally or vertically (state horizontally or vertically (state which required). Tuner can be readily mounted at any angle. $250 / 300$ v. at $20 \mathrm{~m} / \mathrm{a}$. Price $\langle 1 i / 11 / /-$ $250 / 300 \mathrm{v}$. at $20 \mathrm{~m} / \mathrm{a}$.
Carr. and pkg. $5 /$.
No. 2/VS "SYMPHONY" SUPERHET TUNER. As No. 2 but incorporating on the wave change switch an extra position for radio, giving T.R.F. bandwidth Price 13 gns . Carr. and pkg. 5/-

## TAPE DECKS \& AMPLIFIERS

 TRUVOX Tape Deck Mark III T.R.2/U. Latest version to take pre-recorded tapes. Price 22 gns . Illustrated leaflet $2 \frac{1}{\mathrm{t}} \mathrm{d}$.TAPE AMPLIFIER TYPE $C$, expressly designed by Truvox to work perfectly with their Deck, 3 valves plus rectiffer and Magic Eye level indicator. Price 16 gns PORTABLE TAPE RECORDER CABINET to house Truvox Tape Deck and Amplifier together with speaker. Very strongly made and attractively finished in Rexine. Price $64 / 15 /$ - carriage paid.
BRENELLL TAPE DECK
BRENELL TAPE DECK
This new Tape Deck has provision for 3 i in . 7 tin . and 15 in . per second and has instantaneous braking On 33 i.p.s. it plays for two hours. The heads are designed for High Fidelity Recording and Play-back of music and commercial preImmediate Delivery. Illustrated leaflet $2 \frac{1}{2} d$.
N.R.S. High Fidelity TAPE AMPLIFIER to suit price $16 \frac{1}{\mathrm{~g} n \mathrm{~ns}}$ COMPLETE RECORDER ready
to play. Price 45 gns .
NEW MODEL PORTABLE
RECORD PLAYERS
We are pleased to announce the entry on to the market of two Symphony" Record Players deslgned to represent the greatest value in this line ever offered Model No. I contains the Collaro 3 -speed single record playing unit $\mathrm{AC3} / 554$ and model No. 2 contains she Collaro Autochanger RC54. They are available with
either 'Type " $O$ " insert, " $p$ " insert or transcription insert Prices (in attractive rexine case) No. 1 c $10 / 19 / 6$. No. 2 £ $14 / 19 / 6$ Carr. 7/6. Transcription insert 6/9 extra.


## ATUMSMN is the name that Pye Limited

 have given to their new military equipment, the W.S. C.12. They have chosen this title to readily convey the role for which the set has been designed. Nuclear warfare demands maximum deployment over large areas with excellent communication facilities between vehicles, so that complete co-ordination can be achieved without delay. In the event of atomic attack, vehicular communications must span the area of devastation.


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## A NEW TECHNIQUE IN HIGH SPEED WAVEFORM MONITORING

BANDWIDTH:
$10 \mathrm{kc} / \mathrm{s}$ to $300 \cdot \mathrm{mc} / \mathrm{s}$
INPUT IMPEDANCE OF EACH PROBE:
Approx. 1 pf (input element of variable capacity divider)

MAXIMUM SENSITIVITY:
Full Scale Deflection for 1 Volt input
TIME SCALE:
Variable from .05 microsecs to 5 microsecs RECURRENCE RATE OF MONITORED WAVEFORM:
$100 \mathrm{c} / \mathrm{s}$ to $10 \mathrm{kc} / \mathrm{s}$.
CALIBRATION:
Provision is made for accurate measurement of time and voltage scales of a waveform
PREVENTION OF JITTER:
A circuit is incorporated for providing a stable display when a monitored waveform is jittering with respect to its driving pulse.

## MIGH SPEED RECURRENT WAVEFORM MONITOR TYPE 500

The wide bandwidth and high sensitivity of the instrument as well as the very high input impedance result from the use of a sampling technique.
During each recurrence a measurement is made of the instantaneous amplitude of one point in the waveform. This measurement is amplified and applied to the cathode ray tube as one co-ordinate of a graph of the waveform. During subsequent recurrences, instantaneous measurements are made of different points, resulting, after about 100 recurrences, in a complete graph

Member of the AEI group of companies


2-3 kW/Channelised Transmitter

The GFT. $560 / 2$ is a $2-3 \mathrm{~kW}$ channelised transmitter with a frequency range of $1.5-30$ $\mathrm{Mc} / \mathrm{s}$. It consists of three basic cabinets-r.f. unit, modulator unit, and power supply unit-combinations of which can be used to provide multi-frequency working as well as a number of different types of emission. The wave change facilities of the transmitter are both rapid and reliable-a valuable asset when the operating frequency is changed many times each day. The GFT.560/2 is fully tropicalised, and its unit construction facilitates future expansion of the initial installation, should the need arise.
For use in conjunction with the GFT.560/2 there are ancillary units that enable the transmitter to be remotely controlled over a two wire telephone circuit: operational adjustments are dialled to the transmitter.
The versatility and reliability of this new Mullard transmitter make it particularly suitable for h.f. en-route, ground-to-air services and point-to-point communication networks. A team of Mullard communication engineers is available to advise on the use of the GFT. $560 / 2$ in such applications.

ABRIDGED DATA Frequency Range $1.5-30 \mathrm{Mc} / \mathrm{S}$ Frequency Stability To Atlantic City 1947 standards Power Ou:put 3kW. c.w., 2 kW m.c.w. or r/t Types of Emission C.w., m.c.w., telephony, frequency shift A,1, A2, A3, F1 Output Impedance 600 ohms balanced twin feeder Power Supply $400 \mathrm{~V}, 50-60 \mathrm{c} / \mathrm{s}, 3$-phase.



Prompt Delivery, all types, 2,000-20,000 kc/s When ordering ioX replacements, why not use our hermetically sealed Type $2 X L$ ?

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Features that the enthusiast will appreciate are the suppression of switch clicks, the extra heavy balanced curntable, and the very fine degrees of speed control available. Each of the nominal speeds, 78,45 and $33 \frac{1}{3}$ r.p.m. can be adjusted by spproximately $2 \frac{1}{2} \%$. Wow and Flutter have been reduced to less than $0.2 \%$ and less than $0.05 \%$ respectivaly. The model is equipped for dual voltage ranges of 100 to 130 and 200 to 250 volts, 50 or 60 cycles according to the motor pulley fitted. The 301 is finished in quality eray cone namel,
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## AMPLIFIER DESIGNED

AND CUSTOM BUILT IN

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## DISCRIMINATING MUSIC

## LOVERS




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## MAIN AMPLIFIER

Output. 12 watts rated. Peak in excess of 20 watts over $20-25,000 \mathrm{c} / \mathrm{s}$.
Distortion. Total harmonic less than . $1 \%$ at 10 watts- 700 cycles.
Noise Level. 85 DB below rated output. Damping Factor. 50-also variable damping factor through positive to negative values.
Frequency Response. Within 0.2 DB $20-25,000 \mathrm{c} / \mathrm{s} . \pm 0.5 \mathrm{DB} 10-60000 \mathrm{c} / \mathrm{s}$. Feedback. 40 DB total.
Output Impedances. 4 ohms, 7 ohms. 15 ohms.
Input Voltage. 1.2 v for rated output.
Ancillary Power Supplies. 375 volts 30 milliamps, 6.3 volts 3 amps available for VHF Tuner. Pre-amplifier and Tape Reproducer amplifier.
Power Consumption. 130 watts at full load. AC Input 100/150 and 200/250 volts. PRE-AMPLIFIER
Inputs-Magnetic Pickup
B.78. 16 mv input for rated output. $300 \mathrm{c} / \mathrm{s}$ Turnover. Flat above $500 \mathrm{c} / \mathrm{s}$.
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L.P. 13.5 mv input for rated output. $500 \mathrm{c} / \mathrm{s}$ Turnover. 12 DB Roll-off at $10 \mathrm{~K} \mathrm{c/s}. \mathrm{Flat-}$ tened LF at $50 \mathrm{c} / \mathrm{s}$ to +13 DB .
R.I.A.A. 11.5 mv input for rated output. 500 $\mathrm{c} / \mathrm{s}$ Turnover. 14 DB Roll-off at $10 \mathrm{~K} \mathrm{c/s} .3$ DB Flattening at $50 \mathrm{c} / \mathrm{s}$ to +16 DB. at 30 $\mathrm{c} / \mathrm{s}+18.5 \mathrm{DB}$.
Crystal Pickub 35 volt, with inbuilt equalisation from constant amplitude output to constant velocity output enabling switched replaying characteristics to be accurately employed.
Radio/Tape High Level 200 mv. Flat characteristic. Low Level 50 mı. Flat characteristic.
Microphone 6.5 mv for rated output. Flat characteristic.
Mixer Facilities for microphone input] with radio/tape/gramo inputs.
Output. 1.2 voles from cathode follower stage.
Tape Recording Output. 1.2 volts cathode follower independent of monitoring.
Bass \& Treble. Plus and minus 14 DB at $50 \mathrm{c} / \mathrm{s}$ and $10,000 \mathrm{c} / \mathrm{s}$.
Volume. Twin ganged control giving correct gradation.
Low-Pass Filter. Switched $10 \mathrm{Kcs}, 7 \mathrm{Kcs}$, 5 Kcs , and Flat.
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Filter Slope. Variable to 35 DB per octave.
Power Requirements. $375 \mathrm{v} / 7 \mathrm{ma}$.
$6.3 \mathrm{v} / \mathrm{I} \mathrm{amp}$.
Price 44800 complete


* 12 watts undistorted output
$\star$ Infinite damping factor

* Frequency response substantially flat from 2 to 160,000 cycles
$\star$ Ease of mounting
$\star$ Handles records, tape, radio and microphone


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$13 \frac{1_{2}^{\prime \prime}}{} \times 10^{\prime \prime} \times 7^{\prime \prime} .34 .5 \times 25.4 \times 17.7 \mathrm{cms}$.
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$10 \frac{3}{4 \prime \prime} \times 3 \frac{7}{8}{ }^{\prime \prime} \times 4^{\prime \prime} .25 .6 \times 9.2 \times 10.2 \mathrm{cms}$.

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Six positions-tape, radio, microphons and correction for all types of records.

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10 indicated stages (5 cut 5 boost), over 27 dbs. range. No distortion.

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Cuts treble at 4,000 , 7,000 and 12,000 cycles. This non-distorting device enables surface noise, high frequency interference, heterodyne whistle, etc., to be reduced.


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[^13]

## PAINTON WINKLER SWITCH

VOLTAGE RATING: 250 volts A.C./D.C. (maximum).
CURRENT RATING : 0.5 amp . (maximum).
Switching up to 29 positions (single-pole) per bank, or up to 30 positions per bank for $360^{\circ}$ rotation. SINGLE, DOUBLE, THREE-POLE or FOUR-POLE. Painton Winker Switches can be supplied for either 'Make-before-Break' or 'Break-beforeMake ' operation.

## 1-6 BANKS OPERATEDFROMA COMMON SHAFT.

Each switch has an adjustable stop device, by which the switch can be set to the number of positions required.

## AVERAGE CONTACT RESISTANCE : BETTER THAN 0.004 OHMS.

The distinctive Painton.knob type K21, with the 'adjustable skirt' feature has been specially
designed to operate Painton Winkle Switches.

## PAINTON <br> PAINTON Noithamption England




The white pointer can easily be lIned up with dial markings. The friction-plate can be loosened by two screws, allowing the skirt of the knob to rotate.

## for the most uniform response

Of all the different bases that are used for magnetic recording tapes, none can match the precise uniformity of cast cellulose acetate. 'Scotch Boy III', with its cellulose acetate base, offers recordists the most exact uniformity of response that any tape can provide. 'Scotch Boy IIr' is the best of all tapes for high-precision recording, whether of voice, instrument, or mechanical sound.

For laboratory experiments that require the utmost uniformity of response 'Scotch Boy III' is the natural choice: at mooo $\mathrm{c} / \mathrm{s}$ its


Phow and osctllograph of Cy Laurie playing a characteristically apile embroidery of a phrase from "King of the Zulus".
output variation within each reel is less than $\pm \frac{1}{4 b}$., and the variation from reel to reel is less than $\frac{1}{2} \mathrm{db}$. 'Scotch Boy iry' is used by the services for experiments that involve the precise measurement of mechanical and other sounds, and by sound technicians and expert recordists all over the world.
'Scotch Boy 111' is supplied in $1200-\mathrm{ft}$. lenoths on easily-threaded. 7 plastic spools, and also in 600 -ft. and $2400-\mathrm{ft}$. lengths. All these lengths are free from splices.


Record on 'Scotch Boy III'
-the tape with the cellulose acetate base 'SCOTCH BOY'
magnetic recording tape another 3M Product

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The EW54 p-n junction germanium diode is intended for use in rectifier circuits at medium voltage and current. The diode is hermetically sealed in a copper container. This is particularly important because of the deleterious effects of moisture on germanium devices.

The main features of this type of diode are high rectification efficiency and small size; the former results primarily from the very low forward resistance of the diode.

The diode is of value in applications requiring outputs up to the order of 20 A at voltages up to 50 (at $20^{\circ} \mathrm{C}$ ), using a full-wave bridge arrangement.


Details of all the above devices may be obtained from
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> RONETTE developed the type 44 microphone especially for use with portable tape recorders. The plastic housing in "American" style has a loaded base. It can be supplied with any of our various inserts, but it is regularly supplied with the type DX- 12 . On quantity orders the RONETTE badge may be changed for your own trade mark. Fitted with $61 / 2$ ft screened cable.

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Size only $1 \frac{1}{4} \times 3 \frac{1}{2} \times 2 \frac{1}{2}$ with variable iron-dust cores and Polystyrene formers. Built-in trimmers. Tropicalised. Prealigned Retrimmers. Tropicalised. Prealigned Re-
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## The NEW Osmor

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Complete and Prealigned full circuit included. State which station required. 2 M .W., 1 L . W. or $3 \mathrm{MoW} .48 /=$ Inc. P. Tax.

SUPER ' $Q$ ' CUP COILS For Maximum Selectivity A full range is available for all popular wavebands and purposes. The magnetic screening of the cup prevents other components from absorbing the coil's power, thus maintaining the high " Q " value. Simple one-hole fixing. $t$ Only Iin. high. $*$ Packed in damp proof containers. * Adjustable iron-dust cores. $\star$ Fitted tass for easy
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Send 5d. (stamps) for fully deseriptive literature including Circuit and practical Drawings of F.M. TUNER, "The really efficient 5 -valve Superhet," 6 -valve s'het., 3 -valve (plus rectifier) T.R.F. circuit. Battery portable Superhet circuit, Coil and Coilpack leaflets, and full radio and component lists, and interesting miniature circuits, etc.

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I wish to convert my set to pre-set runing. Can this be done fairly easily ?
It is quite simple to convert the standard variable tuned receiver to pre-set. Generally, this means sevitching in various fixed capacitors in place of (and also) the existing 2-gang. Please enquire for lisz of values required for all Stations.

## Dear Sirs,

Please explain the best way to connect an external aerial to my portable s'het.

The outside aerial may be connected at either end of the F.A. or inductively coupled by several turns of the lead-in (not actually connected).


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## New beam tetrode D.C. control valve of exceptional performance

## gm = 35mA/V

## Max. cathode current $\mathbf{8 0 0 m A}$

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In either of these functions the 13.E. 1 can usually be used in place of two or three smaller valves thereby saving space and simplifying wiring because multiplicity of connections, grid and anode stopper resistors etc., are avoided, and this, in turn, gives the additional advantage of improved circuit stability.

The 13.E. 1 has a B.7A. all glass base and is intended for vertical mounting. All maximum ratings shown below are absolute values, not design centres.


THE EDISON SWAN ELECTRIC CO. LTD. 155 Charing Cross Road, London, W.C.2. Telephone: Gerrard 8660.

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Technically identical. with the world-famous Deck supplied, in bulk, to Recorder Manufacturers. With B.S.S. sense of tracking, it is fully approved for playback of pre-recorded tapes. List Price remains at 22 gns.

Details of complete recorders incorporating the TRUVOX Tape Deck are avallable on request.

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These connectors enable identical cables to be joined together, or flexible cables to be joined to their lead-sheathed counterparts. They are fully waterproofed and suitable for use up to $3,000 \mathrm{Mc} / \mathrm{s}$. The flange may be fitted to either half for passing cables through parels.
Illustrated is the $53 / 29 \mathrm{M}$ Plug and $53 \mathrm{~S} / 29 \mathrm{M}$ socket, for use with PT29M cable; also the 53C panel-mounting socket for terminating cable runs and the 53S protective cap for excluding dust and moisture.


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:cables

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 are precision miniature connectors, designed for instrument and aircraft applications where space is limited.Plug and socket bodies moulded in Nylon Loaded P.F.


Actual size
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Contacts finished in pure gold plating to ensure ease of soldering, low contact resistance, and long storage life without corrosion.
Covers made of anodised aluminium fitted with sturdy cable clamp.
The full range, which will shortly be available, will include: $9,18,26$ and 34 way.

[^14]
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## SPECIFICATION

## ELECTRICAL

Supplies: 110, 120 and 200 to 250 volts A.C. $50 / 60 \mathrm{c} / \mathrm{s}$. (200-250 volts in 10 V steps) 90 watts. Cathode Ray Tube: $21^{\prime \prime}$ ( 6.35 cms .) diameter, green phosphor, medium persistence. Voltmeter: Centre zero reading with $4^{* \prime}$ scale. Voltage Measurement: From 0.2 volt to 500 volts A.C./D.C., positive or negative, in 5 ranges. Y Amplifier: Single stage push-pull. D.C. coupled. Deflection Sensitifity: $1 \mathrm{~cm} / \mathrm{volt}$. Bandwidth: $1-3 \mathrm{Mc} / \mathrm{s}$. Input Circuit: D.C. or A.C. coupled. Maximum input: 500 volts A.C./D.C., positive or negative. Attenuators $2: 1,10: 1$, frequency compensated, Input Shunt Resistance: 1 megohm. External reference
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## PHYSICAL

Dimensions: Height $9 f^{\prime \prime}$ ( 23 cms ); Width $87^{\prime \prime}$ ( 20.5 cms ) ; Depth $11 \frac{1}{2}$ " ( 29 cms .). Weight: $22 \frac{1}{2} \mathrm{lbs}$. ( 9.80 kgs .). Finish: Silver Grey Hammer.

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## (2)

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[^15]

## You are there...

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Egen Potentiometers are based on long experience of requirements of television and electronic equipment manufacturers. In design, dependability, accuracy and freedom from wear they are outstanding, but, above all, they are completely NOISELESS.

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## FULLY INTERLEAVED

SCREENED AND IMPREGNATED. ALL GUARANTEED
ALL PRIMARIES ARE 200/250 v. Half Shrouded.

| $\text { at } 2$ | 3 |
| :---: | :---: |
| S63. Output $250-0-250$ v. $60 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v}$, at $3 \mathrm{amps} ., 5 \mathrm{v}$. at |  |
| 2 amps......................................................................... | 6/6 |
| S40. Windings as above. 4 v . at 4 amps ., Ouspur. |  |
| 1S2. $250-0-250$ v. $80 \mathrm{~m} / \mathrm{l}$ |  |
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| HS2X. 250-0-250 v. $100 \mathrm{~m} / \mathrm{a} ., 21 /$. HS75. 275-0-275 v. |  |
| S30X. $300-300$ v. $100 \mathrm{~m} / \mathrm{a} ., 21 / \mathrm{l}$. HS3x. 35 |  |
| 100 m |  |

Fully Shrouded
FSM63 (Midget). Output $250-0-250 \mathrm{v} .60 \mathrm{~m} / \mathrm{a}, 6.3 \mathrm{v}$. at $3 \mathrm{amps} .$, 5 v. 2 amps
Output
FS2. $250-0-250$ ч. $80 \mathrm{~m} / \mathrm{a}$.
FS30. $300-0-300$ v. $80 \mathrm{~m} / \mathrm{a}$. $21 / \mathrm{l}$. FS $3.350-0-350 \mathrm{v} .80 \mathrm{~m} / \mathrm{a}$. ...
FS2X. $250-0-250$ v. $100 \mathrm{~m} / \mathrm{a}$. 23/-. FS75. $275-0-275 \mathrm{v} .100 \mathrm{~m} / \mathrm{a}$.
FS30X. $300-0-300 \mathrm{v} .100 \mathrm{~m} / \mathrm{a}$. 23/-. FS3X. 350-0-350 v. $100 \mathrm{~m} / \mathrm{a}$.
All the above have $6.34-0 \mathrm{v}$. at $4 \mathrm{amps}$. , 5-4-0 at 2 amps .
FS43. Output $425-0-425 \mathrm{v} .200 \mathrm{~m} / \mathrm{a}, 6.3 \mathrm{v} .4 \mathrm{amps.}$, C.T. 6.3 v . 4 amps., C.T. 5 v. 3 amps. Fully shrouded ............................... FS50. Output $450-0-450$ v. $250 \mathrm{~m} / \mathrm{a} ., 6.3$ v. 2 amps, C. T. 6.3 v 4 amps. C.T. 5 v. 3 amps. Fully shrouded
F35X. Output $350-0.350$ v. $250 \mathrm{~m} / \mathrm{a} ., 6.3$ v. $6 \mathrm{amps} ., 4 \mathrm{v} .8 \mathrm{amps}$., 4 v. 3 amps., 0-2-6.3 v. 2 amps. Fully shrouded .................... FSI $60 \times$. Output $350-0-350$ v. $160 \mathrm{~m} / \mathrm{a}, 6.3 \mathrm{v} .6 \mathrm{amps.}$,6.3 v .
 6 amps.. 5 v. 3 amps. Fully shrouded
HS6. Output $250-0.250$ v. $100 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v} .6$ amps., C.T. 5 v. 3 amps. For receiver R1355. Halfshrouded ..........................
HS150. Output $350-0-350$ v. $150 \mathrm{~m} / \mathrm{a} ., 6.3$ v. $3 \mathrm{amps} .$, C.T. 5 v. 3 amps. Half shrouded $\ldots \ldots . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . ~$ v. 3 amps. Fully shrouded
FSI20. Output $350-0-350$ v. $120 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v} .2 \mathrm{amps} .$, C.T. 6.3 v . 2 amps., C.T. 5 v. 3 amps. Fully shrouded
PRI/1. Output 230 v. at $30 \mathrm{~m} / \mathrm{a} .6 .3 \mathrm{v}$. at $1.5 / 2 \mathrm{amps}$. ................ FSI50x. Output $350-0-350 \mathrm{v}$. at $150 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v}$. at 2 amps., C.T. 6.3 v . at 2 amps ., C.T. 5 v. at 3 amps. Fully shrouded ...... The above have inputs of 2001250 v .

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All 200/250 v. Input.

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 F29. 0-2-4-5.6.3 v. @ 4 amps., I8/9. FUil2.0-4-6.3 v. @ 3 amps. FU24. 0-12-24 v. @1 amp...
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F6/4. Four windings at 6.3 v. tapped 5 v.@ 5 amps. each, giving
by suitable series and parallel connections up so 6.3 V . (3) 20 amps.

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Published for "Wireless World"

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To meet the exacting demands being made on the efficiency of aerial systems, the Glover range of Cellular Polythene insulated downleads have been designed to utilise the superior electrical properties of this new form of polythene.
Details of three designs are given as being most representative of modern practice.

The two Cables G.R.1., G.R.2. are intended for use in the service area and one G.R.3. for use in fringe areas and in situations where interference is high.

|  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Reference No. | G. R. 1. 5 | G. R. 1, F. | G. R. 2. S. | G. R.2.F. | G. R. 3 . |
| Characteristic Impedance ohms. | 75 | 75 | 75 | 75 | 75 |
| Service Area | LOCAL | LOCAL | LOCAL | LOCAL | FRINGE |
| Attenuation d8/100 ft. at $50 \mathrm{Mc} / \mathrm{s}$. | $3 \cdot 0$ | 3. 4 | $2 \cdot 3$ | 2.6 | 1.5 |
| . ${ }^{\text {c }}$. $200 \mathrm{Mc} / \mathrm{s}_{\text {, }}$ |  |  |  |  |  |
| Copper Conductor | 1/022 | 7/0076 | 1/029 |  | 3. ${ }^{\text {a }}$ |
| Diam in inches:- |  |  |  | 7/010 | 1/044 |
| Over Polythene. | 0.093 | 0.093 | 0.128 | 0.128 | 0.200 |
| , Wire Braid. | 0.117 | 0.117 | 0.152 | 0.152 | 0.200 |
| .. P.V.C. Sheath. | 0-157 | 0.157 | 0.202 | $0 \cdot 202$ | 0.290 |

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## Video Oscillator - Type 0.22B

A portable instrument covering the range $10 \mathrm{Kc} / \mathrm{s}$ to $10 \mathrm{Mc} / \mathrm{s}$ with an output of +10 dbs to - 50 dbs on 1 volt p. to p. amplitude stabilised to 0.5 dbs over its full frequency range. It includes a 50 cps. square wave output and facilities for direct reading of the modulus of the load impedance.

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A wide range bridge for the measurement of resistance, capacitance and inductance between $15 \mathrm{Kc} / \mathrm{s}$ and $5 \mathrm{Mc} / \mathrm{s}$. It will measure complex impedances balanced and unbalanced and between any pair of terminals in a three terminal network.


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Complete with microphone and tape. super tape deck. Sufficiently powerful to meet any volume requirement. Fitted in padded simulated crocodile case with continental gilt fittings. Hi-Fi model with additional 10 in . speaker built in detachable lid

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S.L. 90

S.L. 86


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[^16]Photograph of a $3 \frac{1}{2}{ }^{\prime \prime}$ Moving Coil D.C. Microammeter by courtesy of The Weir Electrical Instrument Co. Led,

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12 v .1 amp. | $12 \mathrm{v},. 2 \mathrm{amp}$ |
| :--- |
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SP3514， $350-0-360,180$ mA．．．．．．．．．．．．．．．．．．．．．．．．．．． $2-4 \mathrm{~F}$

SP352， $350-0-350,160 \mathrm{~mA} ., 5 \mathrm{v}$ ，© $2-3 \mathrm{a}, 6.3 \mathrm{v}$ ，（5）
SP425A，4．8 $426 \cdot 0-425,200 \mathrm{~mA} ., 6.3 \mathrm{v}$ ，（⿴囗十）2－3 a．， 6.3 v ．
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 $200-230-250$ output $3 \nabla .-30 \vee$ ．，© 2 a ． $17 / 6$
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AcOs．High impedance Cryatal Microphone，type 35－1， ACO5．－High Impedance Crystal Microphone，type 33－1， ACOS．＂Mic30＂impedance Cryatal Mierophone $22 / 10 \%$ ． （Thia Microptone can be used as eitber Hand or Deatk type．）

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#  

## 32. A HIGH GAIN VIDEO AMPLIFIER

The gain of a video amplifier with a given bandwidth, using a high slope r.f. pentode, is limited by the output or load capacitance. A typical cathode compensated circuit using an EF80, with a bandwidth of $3 \mathrm{Mc} / \mathrm{s}(6 \mathrm{~dB})$, a maximum video output voltage of 50 V , and a load capacitance of 26 pF , has a gain of 11 . The gain can be increased and the bandwidth maintained if the effective load capacitance is reduced by means of a cathode follower. The practical circuit shown in Fig. 1 has a gain of 22. A Mullard PCF80 triode pentode combines the functions of video amplifier and cathode follower in one envelope.

FIG I


## Video Amplifier

The cathode of the pentode (Fig. 1) is held at about +3.5 V by the pentode cathode current and the bleed current through the $27 \mathrm{k} \Omega$ resistor. With the screen grid at 190 V this gives an anode current of 3.5 mA . The permissible anode current swing is limited by the maximum screen grid dissipation rating of 0.5 W , therefore the output voltage is limited to about 80 V . The dynamic characteristic of a typical PCF80 pentode shows that the corresponding maximum limit of the anode current is 8.8 mA . The effective slope is $1.5 \mathrm{~mA} / \mathrm{V}$, therefore the gain $\mathrm{gm}_{\mathrm{m}} \mathbf{R}_{1}=$ $1.5 \mathrm{~mA} / \mathrm{V} \times 15 \mathrm{k} \Omega=22.5$.

## Cathode Follower

The total capacitive load of a video stage is made up, in a typical good design, approximately as follows: Video output 4 pF , C.R.T. 8 pF , Synchronising separator 6 pF , Noise suppressor 2 pF , Strays $6 \mathrm{pF}=26 \mathrm{pF}$.

If the total gain of the stage is to be greater than 20 and a cathode follower output (with a gain less than 1) is to be used, the amplifier must have a higher gain-say 25 . To achieve this the output capacitance must be less than 11 pF . With a cathode follower the capacitive load of the amplifier is reduced to: Video output 4 pF , Cathode follower input 3 pF , Strays $3 \mathrm{pF}=10 \mathrm{pF}$, which is within the 11 pF limit. The overall response is limited by the cathode follower, which must be capable of developing an output voltage of at least 50 V across a capacitive load of 26 pF .

Further notes on the circuit are available with reprints of this advertisement. Reprints of all advertisements in the Valves, Tubes and Circuits series may be obtained free from the address below.
mULLARD LTD., Technical Service Dept., Century House, Shaftesbury Avenue, London, W.C. 2
MVM 327


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A crystal hand ordeskomnidirectional microphone for the high quality public address and tape recording field, incorporating a specially designed acoustic filter giving a response flat from 30 to $7,000 \mathrm{c} / \mathrm{s}$. RETAIL PRICE: $\mathrm{f}^{2-\mathrm{ro}-\mathrm{od} .}$


A handsome omni-directional instrument of high sensitivity and a substantially flat response from 30 to $7,000 \mathrm{c} / \mathrm{s}$. Alternative models, with or withoutswitch, are a vailable with suitable adaptors for floor or table stands or for hand use.
RETAIL PRICE: E3-3-od. without switch or $£ 3-8-$ od. with switch.

A general purpose hand microphone of robust construction with substantially flat response from 50 to $5,000 \mathrm{c} / \mathrm{s}$. Suitable for recording apparatus. Public Address equipment etc. RETAIL PRICE £I-s-od.
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We shall be at
THE RADIO SHOW
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## "BELLING-LEE" NOTES

AERIALS FOR THE V.H.F. BROADCAST BAND II FREQUENCY MODULATED


The period of experimental B.B.C. transmissions from Wrotham has passed.
From now onwards the Wrotham transmitter will send out three programmes, the Home, the Light and the Third, on Band II. These transmissions will be frequency modulated and will require horizontal aerials which are in the position of maximum response when broadside on to the transmitter. The B.B.C. V.H.F. programmes will eventually practically cover the country, and this will allow listeners to enjoy a quality of reception hitherto unobtainable. The next transmitters to come into operation will be Pontop Pike (Newcastle) and Divis (Belfast) in that order. In many difficult locations listening will become a pleasure for the first time. New or modified receivers will be required and these will have built-in general purpose aerials. However, tentative field trials indicate that, in many cases, and certainly at distances in excess of 40 miles from the transmitter, particularly in hilly country, a better aerial will be required. In a number of cases a "Lofrod" loft aerial will be satisfactory but at these distances many listeners will require an outdoor aerial and in really difficult locations a 3-element or an horizontal "H" will be necessary. Above we illustrate V.H.F. stubs attached to a dipole. These may also be fixed to any of our aerials of the parasitic element type.

> Advertisement of

GELLING \& LEE LTD. Middx.
Written 20th June, 1955

## DIPLEXER TUNED FILTER

 For use with band III aerials, adaptors, and bandI/band IITaerialsIf a television receiver has a single input socket, and if separate band I and band III aerials are being used, viewers can either put up with the inconvenience of changing over the aerial every time they switch to the other programme or they can terminate both aerials into a diplexer and connect the socket output from this to the receiver inpur. Both signals are available at the receiver, but when the selector switch is tuned to one programme, the receiver does not have to contend with interference picked up on the temporarily unwanted aerial. If the feeders from the two aerials are just bunched and connected to the receiver input, there might be ghosting interference from both aerials instead of from only one, impaired signal strength due to mis-match, troubles due to standing waves on the feeders, or other interaction.
1 A diplexer is necessary where separate band I and band III aerials are fed to a receiver having a single input for both frequencies.

2 A diplexer is not needed where separate band I and band III aerials are fed to a receiver having separate input sockets.

3 A diplexer is only required with combined band I/band III aerials, or aerials fitted with adaptor units, where the receiver has separate inputs.

4 On long cable runs, it is often more economical to fit a diplexer at each end, utilising a single feeder low-loss cable.



## COMPLETE

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## ANNOUNCEMENT fM Receiver alignment generator model 1324

This Alignment Generator will be available next month to provide the Service Engineer with a compact test set with which all essential alignment procedures on FM Broadcast Receivers may be undertaken.
Accurate trimming for correct overall and IF response curves is easily carried out and facilities will be provided for discriminator alignment and checks on its sensitivity and distortion.

COSSOR Model 1322

Telecheck and Marker Generator for Bands I and III

Model 1322-used in conjunction with a cathode ray oscillograph-provides equipment for the display, measurement and correct adjustment of RF and IF response curves of television receivers. This entirely new instrument comprises a swept oscillator covering the Television BANDS I and III ( $5-75 \mathrm{Mc} / \mathrm{s}$. and $155-225 \mathrm{Mc} / \mathrm{s}$.) and a frequency marker oscillator so that precise calibration of the oscillograph display may be made; accuracy of the frequency of the marker pips being verified by reference to an internal crystal. The
alignment oscillator is set to the video carrier to which the receiver is tuned and the sweep (either $1 \mathrm{Mc} / \mathrm{s}$. or $10 \mathrm{Mc} / \mathrm{s}$.) is automatically derived from the time base voltage of the display oscillograph. The response of the "strip " under test to the frequency band applied is then presented on the screen of the cathode ray tube. The RF output of Model 1322 is available at 80 ohms and is adjustable from a maximum of 40 millivolts to a minimum of 25 microvolts through a coarse and fine attenuator.

TELECHECK CONVERTER FOR BAND III
Model I32I

This adaptor provides owners of Model 1320 " Telecheck" with an extension of the frequency range of the original instrument into the BAND III television channel. Thus, alignment procedures adopted for BAND I RF/IF "strips " are available also for BAND III receivers. A selection of the desired BAND is made by means of a switch. Pattera generator facilities for picture time base linearity checks have been retained. Model 1321 Adaptor is designed for permanent attachment to the standard "Telecheck" providing a neat, light and compact unit. Mounting is effected by four screws and the inter-connecting wiring is carried in a single insulating sleeve.


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## SUBJECT(S) OF INTEREST

(We shall not worry you with personal visits) AUGUST

IC38:3

## Marconi Double Diversity Telegraph Recording Unit

## TYPE HU. 12

The HU.i2 unit is designed for use with two suitable receivers of frequency shift or on-off telegraph transmissions in double diversity. The combined AF outputs of the receivers are converted into double current signals to operate an undulator, relay or other 'space-mark' type of equipment requiring a current up to 30 mA . No receiver modification is necessary except, perhaps, adjustment of the BFO frequency.
Satisfactory recording is possible should the first oscillator drift up to $\pm 800 \mathrm{c} / \mathrm{s}$.

## FEATURES:-

- Electronic path diversity selector for FSK will function when difference between paths exceeds 4 db .
- Alternative DC output suitable for a low resistance undulator ( $20+20 \mathrm{ohms}$ ) is provided.
- Neon indicators for visual tuning of FSK reception.
- Front of panel control for pre-set 'mark' and 'space' current adjustment, filter selection and signal bias.

Over 80 countries now have Marconi equipped telegraph and communication systems. Many of these are still giving trouble free service after more than 20 years in operation.


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# TAPE RECORDER 



The amplifier, speaker and case, with detachable lid, measures $8 \frac{1}{4} \mathrm{in} . \times 22 \frac{1}{2} \mathrm{in} . \times 15 \frac{3}{4} \mathrm{in}$. and weighs 30 lb .

PRICE, complete with WEARITE TAPE DECK
¢84 00
$\star$ The total hum and noise at $7 \frac{1}{2}$ inches per second $50-12,000 \mathrm{c}$. p.s. unweighted is better than 50 dbs .
$\star$ The meter fitted for reading signal level will also read bias voltage to enable a level response to be obtained under all circumstances. A control is provided for bias adjustment to compensate low mains or ageing values.
$\star$ A lower bias lifts the treble response and increases distortion. A high bias attenuates the treble and reduces distortion. The normal setting is inscribed for each instrument.
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$\star$ A heavy mu-metal shielded microphone transformer is built in for $15-30$ ohms balanced and screened line, and requires only 7 micro-volts approximately to fully load. This is equivalent to 20 ft . from a ribbon microphone and the cable may be extended 440 yds. without appreciable loss.

* The .5 megohm input is fully loaded by 18 millivolts and is suitable for crystal P.U.s, microphone or radio inputs.
* A power plug is provided for a radio feeder unit, etc. Variable bass and treble controls are fitted for control of the play back signal.
* The power output is 3.5 watts heavily damped by negative feed back and an oval internal speaker is built in for monitoring purposes.
* The play back amplifier may be used as a microphone or gramophone amplifier separately or whilst recording is being made. t The unit may be left running on record or play back, even with $1,750 \mathrm{ft}$. reels, with the lid closed.


## POWER SUPPLY UNIT to work from 12 volt Battery wit <br> 50 cycles within I\%. Suppressed for use with Tape Recorder. PRICE \&I8. 00. <br> We supply and recommend the Jason F.M. Feeder Unit. PRICE 1517 0, including Purchase Tax. <br> FOUR CHANNEL BLECTRONIC MIXER

is almost essential for the professional or semiprofessional where a number of different items have to be mixed on one tape recording.
It is recommended by a number of tape recorder manufacturers for this purpose.
Any normal input impedance can be supplied to order, balanced or unbalanced, the standard being 15-30 ohms balanced.
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Used in many hundreds of large public address installations and recording studios throughout the world.


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## THE

## DYNAMIC PICKUP

This pickup will very probably earn recognition as the best in the world. It is a definite advance on the pickup introduced five years ago, which is used in many recording studios for dubbing and playback.

## PRICES

The arm: £2/15/-, plus $19 / 3$ Purchase Tax.
L.P. head with diamond stylus :
£5/15/-, plus £2/0/3 Purchase Tax.
78 head with diamond stylus: £5/15/-, plus £2/0/3 Purchase Tax.
Mumetal-cased transformer : $£ 1 / 15 /-$.

## SPECIFICATION

## * THE ARM

This is of advanced design having very low inertia. Friction is kept to a minimum by using a single pivot bearing. The arm is counter-weighted and has provision for plug-in interchangeable heads. An arm-rest is provided.

## * GENERATING SYSTEM

Dynamic (moving-coil). Coil impedance approximately 6 ohms, $1,000 \mathrm{c} / \mathrm{s}$. No magnetic material is embodled in the moving parts, and the pickup is free from the inherent distortion of moving iron (magnetic variable reluctance) types. These distortions are also inherent in those dynamic pickups in which the moving coil is wound on a magnetic core.

## $\star$ STYLUS

Material : Diamond, guaranteed unconditionally not to chip or break. Stylus sizes : L.P. 0.001 in . radius + nothing -0.0001 in . 78, 0.0025 in . radius $\pm 0.0001 \mathrm{in}$.

## * PLAYING WEIGHTS

Between 2 and 3 grammes for L.P.
Between 5 and 6 grammes for 78 .
Automatically adjusted by the weight of the head.

## t RECORD AND STYLUS WEAR

These are lower than on any pickup of which we have cognisance. Diamond has a playing life of approx. 100 times longer than sapphire, and because it will take a higher polish than any other material it therefore causes less record wear.

## * OUTPUT

The shielded step-up transformer delivers an output of 11 mV for each $\mathrm{cm} / \mathrm{sec}$. r.m.s. recorded velocity. This means that an amplifier with a sensituvity of 40 mV at $1,000 \mathrm{c} / \mathrm{s}$ will be easily loaded by the pickup from commercial records.

## t frequency response

Total variation $\pm 1 \mathrm{db} 20,000 \mathrm{c} / \mathrm{s}$ to $40 \mathrm{c} / \mathrm{s}$ with the $L P$ head, including transformer (recorded velocity $1.2 \mathrm{cms} / \mathrm{sec}$. r.m.s. above turnover). Low frequency resonance :
$20 \mathrm{c} / \mathrm{s} \pm 5 \mathrm{c} / \mathrm{s}$ with our very lightweight arm.
High frequency resonance:
0.001 in . radius Vynil, $21,000 \mathrm{c} / \mathrm{s} \pm 2,000 \mathrm{c} / \mathrm{s}$.
0.0025 in . radius on shellac, above $27,000 \mathrm{c} / \mathrm{s}$.

The frequency response does not change with temperature.

## K SIGNAL-TO-HUM-RATIO

It is not possible to specify this important ratio without stipulating the strength of the interfering fields. These fields will, of course, vary according to the installation. However, for the purpose of comparison measurements have been taken under working conditions, i.e. with various pickups mounted normally within inches of the electric turntable motor and within two feet of a power transformer in an amplifier. The results show that the Leak Dynamic Pickup has a lower hum content than any variable reluctance (moving-iron, magnetic) pickup and a very much lower hum content than a single turn moving coil (i.e. "ribbon") pickup. This confirms what would be expected from theoretical considerations.

## * DIMENSIONS

From the centre of the flxing stem to the front of the pickup head, 91 in . From the centre of the fixing stem to the rear of the arm, 2 in. The height of the pickup is adjustable and it can be used with any turntable.

## * MOUNTING

A template of original Leak design is supplled, enabling the pickup to be accurately located on the turntable mounting board. There is a single fixing hole and the stem contains a miniature socket which accepts the plug leading to the transformer (see illustration).

## * TRANSFORMER

The transformer has a step-up ratio of 1.80 and is heavily shielded in mu-metal. The primary lead is terminated in a plug and a shielded secondary lead is supplied.

* Write for illustrated leaflet ' $W$ '.

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## TRANSMITTER 1131

This is a high powered transmitter for operating over the same frequency range as the Receiver 1132, i.o., 70.130 mega. cycles. It is a very bulky transmitter and probably contains around 8300 worth o equipment. As far as we know these have never been used but of course have been in will foed attention before being put into will need attention before being put int £ $37 / 10 /$-, plus carriage.

## R1132

Wo have $a^{3}$ amall quantity of these re. ceivera still available less valves. Their condition unfortunately is not good bu course, contaln a multitude of spar parts. At $30 /$ each they represent real bargain. If not collecting, please inciude 5/- for packing and carriage.

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For remote con trol of D.C. moto between 1 and 3 or 100 y , or 230 y Unused and in first-class condition, complete with metal and Wired glass cover Price El 10 .


EX-ROYAL NAVY SOUND POWERED TELEPHONE These require no batteries, and will go for long perfods without attention. Complete with generawr and sounder which gives high pitched note, easily heard abo any other noise. Also gited with an can be veed instead of the' sounder, or where several telephones are used together will indicate which one is being called. Bize 7 tin. $\times 9$ in. $\times 7$ in ., wall mountling, designed for shlps ${ }^{\circ}$ use but equally auitable for home, office, ware-.
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Items available, all new and unused, are:Frequency Changer, Type 6AC, Ref. No. YBO2700, price £5. Standard G.P.O. deak type instrument with scramblior ${ }^{8} \mathrm{~m}$ witch, compiete wiith lead and function gezerator in wooden box, 15\%-. Junction box with three maltiple relays and cable strins, $35 /$ - Bank of three drop in-


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## 10-CORE CABLE

10 flexible copper conductors well inrulated suitable for mains work. Covered

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500 WATT 1,000 v. (ADJUSTABLE)
The conventional circuitry is emploved throughout and all componentrs are amply proportioned to permit substantlal overloading. At master switch controls the
whole unit and whenever thls is on Whole unit and whenever
currentis suppli id to the rectifer filaments, current is supplied to the rectificr iamenis,
thus keeping them al ways in the emil selve state. The H.T. transionner is suppiled from the primary of the fllament translorrmer, connection being via an on/off
switch and a tapped choke. The on/off switch controls the H.T. and the tapped choke conjunction with its selector switch gives ten
Two directly heated rectidere give a full wave output which
1s smoothed by a 10 Henry choke and 4 mid. condenser. Ableeder resistor connected across the output serves as a dummy load and also discharges the emoothing condenser which otberwise would be a source of danger to users.
The continuous rating of the power pack is 1,000 volte at 500 mllilamp ( 500 watts). But the proportions of the various components are such that 100 per cent. overloading can be allowed for pulse work or other intermittent operations. The size or the power Price; Kit of parta $£ 27 / 10$-, or made up ready $£ 37 / 10$ /-

## 500 WATT $2,000 \mathrm{v}$. (VARIABLE)

The maximum continuous rating of this is 250 milliampa at 2,000 volts. Rectification Is half wave. Specification otherwise as for the variable $000 / 1,000 \mathrm{\nabla}$.
1,000 WATT $2,000 \mathrm{v}$. (VARIABLE)
The contlnuous power ratlag of this is 500 milliamps at 2,000 Folts. But the tapped 120 lb gize $16 \mathrm{kn} \times 131 \mathrm{n} \times 131 \mathrm{n}$. Price $£ 37 / 10 / \mathrm{h}$ in kit form, or made up ready to use £47/10/-.
1,000 WATT $1,000 \mathrm{v}$. (VARIABLE)
The miaximum continuous rating of this is 1 amp . at 1,000 volts. Rectifeation is ful
 Price $£ 37 / 10 /-$ in kit form, or $£ 47 / 10 /-$ made up ready to work.

## FIXED MODELS

Any of the models mentloned above can be supplied without the tapped choke and selector switch. The prices are as follows;
Mixed $500 / 1,000$ v. $£ 22 / 10 /-$ in $k i t$ form, or $£ 30$ made up. Fixed $250 / 2,000 \nabla$. $£ 22 / 10 /$ - In kit form, or $E 30$ made up Fixed $500 / 2,000 \mathrm{v}$. $£ 32 / 10 /-$ in kit form, or $£ 40 \mathrm{made}$ up Fized $1,000 / 1,000 \mathrm{v}$. $£ 32 / 10 /-\ln \mathrm{klt}$ form, or $£ 40$ made up

## CEILING FAN

This model, made by Revo, incorporates a series-wound totally enclosed ball-bearing motor of robust construction and noiseless operation. The fan has a blade diameter of 36 in . and is supplied with 201 in . suspension tabe and celling canopy. All finlehed white cellulose enamel. The voltage working is $230-250$ v. D.C. Revo catalogue number D12288 Price £10:10/-


## SPECIAL PURPOSE VALVES

Triode Type CV 1098-this is a high-power air-cooled triode. Specification of which is as follows; Filament voltage 8.2 v ., flament carrent 35 amps ., anode dissipation 750 watts. Maximum anode poltrage 23 kV .
This valve is vers sultable for R.F. heating at high frequencies and two of these in push-pull under Class $\mathbf{C}$ conditions would have an output of approximately 2 kilowatts. Brand new, still in original shockproof packing, price £15 each.

TETRODE TYPE VT31 This is a high-powered air.
cooled tetrode. 8 pecifcation of cooled tetrode, Spectication of
Which is as followr :-Heater
volta 11.25 , Heater current 8 vimp., maximum anode voltage 8 kV , anode dissipation 250 watts, size approximately 14tin. long and 6 in. across the
Limited quantity only at $£ 4$
each, still in original packing.


## WATCH THESE COLUMNS FOR DETAILS OF <br> VARIOUS OTHER INDUSTRIAL TYPE VALVES.

 ALTERNATIVELY SEND US YOUR ENQUIRIES.
## HIGH CYCLE MOTOR ALTERNATOR

TYPE 1 Has a motor 230 v., 50 cycle single phase 2,800 r.p.m., coupled to a 8 gither output 250 จ., 1,728 cycles at .24 amps. Good condition, with wiring diagram, $£ 3 / 10 /-$ phas 7,6 cartiage.
TYPE 2. Has a motor 230 v. 60 cycle single phase, conpled to an alternator output

## 250 v. 625 cyclea .24 amps . Price $53 / 10 /$-, plus $7 / 6$ carriage <br> -SPECIAL EQUIPMENT SALES

## IMPORTANT NOTE

Owing to the bulkiness of many of the items Hated on these two pages It may not Owing to the bukiness of many of thes, therefore please telephone conflrmation that
be possible to keep stocks at branches, ther be possible to keep stocks at branches, $\begin{aligned} & \text { itere journeying specially to see it. } \\ & \text { the itemally at the branch belor }\end{aligned}$
SPECIAL SALES DEPT., E.P.E. LTD., BOURNE HOUSE,
GROVEROAD, EASTBOURNE.


Famous wartime "c cat's eye" used or seelng in the dark. This is an infra-red Image converter cell with a silver caesium screen which lighte up (like a cathode ray
tube) when the electrons relcased by the infra-red strike ft. It follows that as light from an ordinary lamp is rich in infra-red these cells will work: burglar alarms, counting circults, smoke detectors and the hundred and one ouner devices as will the ampler type of photo cell. Here then is a golden opportunity for some interesting supplied with cells if requested


METERS

2 in. Flush mounting
0.30 raA . moving coil
0.300 mA . moving coil
5.500 mA , moving coil
$5.0-5 \mathrm{~mA}$. moring coll
2in. Flush mounting
2in. Flush mounting
0.2 amp. R.F. thermo.
0.3 amp . R.F. thermo.
0.3 amp . R.F. thermo.
0.6 amp R.F. thermo.
$0-5 \mathrm{~mA}$. moving coll.
$0-3 \mathrm{~mA}$. moving coil
$0-20 \mathrm{amp}$. moving coil.
$0-40 \mathrm{amp}$. moving coil
Hot Wire Amp. Moter
$0-8$ amp. 2 tin. flush
$10 / 6$
$10 / 6$

PYREX AERIAL INSULATORS
Idea! for serlal connections through cabin wails or through pas
ele. Consists of glase dome with threaded rod and terminal ends and metal fixing flange. Price 2/- each


This brass cased plug and socket is extremely robust and ideal for P.A. or ontside work. Idoal also for taking power to unlcs as it ineulates the onds of the wires. Contacts are quite Enitable for used for lighting or power. Price $2 / 6$ per pair.


JUMBO VALVE BASES
Carsmic 4 -pin for transmitting valves
FLEXIBLE COUPLINGS
(0)71) These are some-
ames known as bellow couplings oxtend as well as bend. They are ldeal for jotning shafts whtch are out of alignment and for slug tuning controla where the core has to come is and out. Price 1/9 each.


30 AMP. ROTARY SWITCH
single pole ON/OFF, a very robust switch. made by one of our most famous firms. Will give life-time of service.


## BAND IIII ITEMS

## THIS MONTH'S SPECIAL

## The first of a series of highly efficient low priced aserials to be made

 by Cleveland Electric.This is an array for Band III technically. known as a folded " $v$ " it is highly efficient, and directional and suitable for indoor or outdoor use, even for fitting on the window frame or coping or existing T.v. mast, it employs polythene low loss insulators and non-corrodable elements. and packing 1/8.

PRE-ASSEMBLED WITH ${ }^{3}$ in. DIA. RODS
3-element armay with swanneck mast with ' U' bolt
elamp for fiting to exiting masts from tin . to 2 in. dia... 3 -element array with cranked mast and wall mounting
bracket bracket
3 -element array with cranked matt and chimney lashing 5 equlpment
clamp for array with swanneck mast and ' $U$ ' bo's carap for fitting to existing mast from fin to 2 in . dia... 5 -element array with cranked mast and wall mounting bracket
5.element array with cranked mast and chimpey lashing equipment
8 -element array with swanneck mast and $\cup$ ' bodt clamp for fitting to lin. to zin. dia. mast
8-element arriuy with cranked mast and chimney lashing equipment
8 eelement array with 1 inin. mast cap, loft. mast and heavy duty ingle chimney lasbing e quipment 10-element array with cranked mast and chimney equipment 0-element array with iyin. mast cap, ioft. mast and heavy duty single chimney lashing equipment heavy duty double chimney lashing equipment.

## WHAT IS IT?

It is the Indicator that you would make to check that the "Elpreq Band III Signal Generator " is working properily. When the loop is brought up to the output clrcult the amp lights brightly


THE "ELPREQ" Band III SIGNAL GENERATOR is most useful. It:1. Will provide the signal for tuning to any Band III station.

Can be used as a grid-dip meter for checking the frequency of Band III T.V. serials, Coils, etc.
4. Can be made to give a pattern on T. F. Recelver sereen.

All the parts including valves, tuning condenser and metal chassis are available as a Kit at $25 /-$ post free. Const ructional data free with Kit or availabie se parately price $2 / 6$.


BE PFADY: COMMERCIAL T.V.
Daily Test Transmissions are already taking place
Our convertor which fixes to the side or back of your T.V. will give you the new station or the old by the dick of a switch. You do nothing to your existing net; just
plug in mains and aerial leads. Suitable for any T.W. plug in mains and aerial leads. Suitable for any T.V.
Price $£ 6 / 10 /$, or $30 /-$ deposit and six payments of 1 .

BUILD YOUR OWN CONVERTER The Converter has given very satisfactory results from the experimental Beulah Hill station. It use 2 valves, is not at all diffcult to make and can be lined up with the simple $25 /$ - instrument described above Price for all the componente including constructional data la- $£ 3 / 10 /-$ or $£ 4 / 10$ - with mains equipment. Price includes stove enamelled case, prepared metal chasois, and all parts to make up convertor as Ulus-trated-data available separatels price $2 / 6$

## MADE FOR THE JOB

This ex.W.D. 10 -valve superbet was designed to re. ceive 200 megacycles transmissions so it will require virtually no conversion to receive the commercial T.V. programmes.
These contain 6 valves type sP61, and one each RL7. BL16 and EA50. S1x IF transformers $12 \mathrm{Me} / \mathrm{s}$, band and hundreds of other use ful components. Price 59/6, plus carriage and packing 7/6 These receivers are unused and perfect

## In the ELPREQ F.M. Tuner tour valves and two cryatals are used. The last valve

 arts as a limiter so reducing the necessity of eract tuning and at the same time heater-cath ode hum so often encountered with valve ratio detectors, Stab ility io oxtremely good and tuning most simple. The tuner draws its power supplied from the set or amplifler, its valve heaters are not connected to earth.Whit only a simple indoor ae rial made by parting the ends of ordinary flexible cable this tuner works very well at East bourne (over 60 milles 1 rom Londos) and we await reports from even greater distances.
Cost of all parts including valvcs, prepared metal chassis, scale, slow motion drive, pointer, tuming knob, in fact evervthing needed to make the completc unit suitable a separate cabinet, is $£ 8 / 12 / 6$, data is included free with the parts or is available зeparately price $2 /-$.

SALE! special prices this month


CONNECTING WIRE
P.V.C. covered in 300 ft . colls-most colours-ifour colls, different colours, $9 /=$

## 3-SPEED RECORD PLAYER



3-gpeed record player with pick-up using the famous Acos "Hi G"turnover crystal -motor ako by very famous maker. All on unit loard ready for installation. A wonderful bargsin at $£ 5 / 10 /-$, plus $5 /$ : carriage.


Carr. $3 / 6$.
 TELECABINET
Veneered and Pollshed-Perfect. New and unused.

## CARBON RESISTORS

50 assorted $t$ and \& watt resistors. Ranging between 10 obms and 10 meg. ohms. (Our 7/6.

## $12^{\prime} 6$



CONSTRUCTOR'S PARCEL
Five valve guperhet chassis-size $15 \times$ $5 \times 2 \mathrm{in}$. With three waveband glass scale, Also TABLE CABINET eneered rad polished
chassis, 38/6. Post $2 / 6$.

NOVELTY RADIO
Complete tunable M/L Radlo with room for 3 in. neaker in base. Needs only vaives, speaker and batteries, 28/6, plus, 2/6. postage, etc.

BEETHOVEN CHASSIS
 ansh-with-order price this month £5/19/6, carriage and insurance 7/6.


BAKELITE CABINET
Two tone with built-in handle, Note-All have slight imperfections but these are soundness of the set


INSTRUMENT CASE
Veneered and polished-undrilled.


Bakelite cabinet. complete with dial, metal chassis and back, and plans of

1/= $\underset{\text { for } 4 .}{\text { Post }}$

UNBREAKABLE GLASS
PANELS


Size 10 a $\times 9$ tin.-parcel of five panela 5/-. Post free.


## CORNER CONSOLE

A massive cabinct but being corner fitted is not out of place even in a modern small living room. Overall dimensions of this cabinet are 47in, wide $\times 31$ in. (deep to corner) $\times$ 50 in . high. Made to house 15 in . Televisor, Radio Unit, Amplifier, Tape Deck, etc. Originally $£ 18$. Our Price- $£ 10$ plus 30/- carriage.

## -THIS MONTH'S SNIP Prepared metal chassis for 5 -valve superhet size approximately $18^{*} \times 81^{\circ} \times 3^{\circ}$ switch, and scale. <br> These are ex-equipment, Iully reconditioned tested and guaranteed. Price is ponly $14 / 6$ (less than the value of coil pack only), complete with circuit diagram,

## SELECTIONS FROM MISCELLANEOUS STOCKS

## TRANSORMER PRIMARIES

 These are tranaformers with wound, primaries tapped 200,220 and 240 but with no secondarles. There is ample window ppace, however, for the handwinding of secondary to sait your own wiading of secondary
roquirements. Number of turns per volt required, depends upon the overall size of the stack, for example, our 100 turns per volt. The size required depends upon amps. taken out and the voltage, eg. if 10 amps, at 10 volts is required then a 100 -watt primary is needed. 50 Amps. at 5 volts requires 250 watt,
etc., etc. Tranaformers are complete etc., etc. Transformers are complete

| clamps <br> Watts | feet. |  |
| :---: | :---: | :---: |
| 80 | Price | Post \& Pks. |
| 100 | $10 / 6$ | $1 / 6$ |
| 100 | $12 / 6$ | $2 /-$ |
| 250 | $12 / 6$ | $2 / 6$ |
| 500 | $25 /-$ | $4 / 6$ |
| 1,000 | $40 /-$ | $6 / 6$ |
| 2,000 | $00 /-$ | $8 / 6$ |



Stud Switch for bullding up variable resintors as illust:ated or similar devices, complete with handle, metal side pieces, and apucer bolts. Price 6/6, post und packing $1 /$
SPECIAL BARGAINS FOR CALLERS at all BRANCHES

## SELECTIONS FROM OUR RANGE OF CABINETS



## THE CONTEMPORARY

Also in the modern trend is this very stylish contemporary console. Veneered in oak with con trasting mouldings, and is ideal for use with modern furniture or with other contemporary fittings or furnishings. The radio and motor board is uncut and its size, $30 \times 15 \frac{1}{2}$ in., provides ample room for all equipment. Price £8/15/-, carriage etc., $12 / 6$.
 EMPRESS CONSOLE
This cabinet is undoubtedly a beautiful piece of furniture. It is elegantly veneered externally in figured walnut, internally in white sycamore. The radio section is raised to convenient level but is not drilled or cut. The lower deck acts as the motor board, again is uncut, it measures $16 \times 14 \mathrm{in}$. and has a clearance of 5 in . from the lid. There is a compartment for the storage of recordings. Overall dimensions of this essentially modern cabinet are 3 ft . wide, 2 ft . 8 in . high, and lft. $4 \frac{1}{2}$ in. deep. 8 in. high, and
Price $£ 15 / 15 /=$, carriage etc., $12 / 6$


OFFERED AT APPROX. HALF COST TO MAKE An impressive costly looking cabinet originally designed for T.V. but simple modification makes the cabinet suitable for radiogram, amplifier, tape recorder, or reflex speaker-size 23 in , wide, 22 in . deep and $37 \frac{1}{2} \mathrm{in}$. high. Limited quanity left at $£ 8 / 15 / \mathrm{m}$ each, which is approximately half of their manufacturing mately half of their manufacturing cost. Also slightly damaged but re-

## BEST SELECTION OF CHASSIS IN LONDON

## THE

## CLEVELAND "ORGANTONE"

5-Valve 3-wave band superhet covering long, medium and short wave
Osram miniature valves are employed and low loss iron cored coils account for an excellent mignal-to-noiso ratio. frequency changer and I.F.
The output stage utiliges vari-
able negative feedback. A gram. position is provided and reproduction of record is particulariy good.
Chasgis size is $12 \times 7 \times 71 \mathrm{n}$.-scale sizies is $10 \$ \times 4 \mathrm{fm}$.
This reciver has been tested in particularly difticult areas and its stability and nosse rejection hive produced exceptional results.
rod aerials $£ 12 / 10 /-$. Hire purchase deposit $£ 4{ }^{7 / 6}$ Slmilar model but with territe

## THE LATEST

DULCI
This ts the Model F3PP. Developed especially to mect the Increasing
demand for high fldelity equipment dearticularly zuitable for reqlacement in a radiogratn. This is a 7 -valve 3 wave band superhet with push-pull output, lucorporating separate bass and treble controls thereby ensuring a maximum control of fldellity volume and tone. Wave band
coverage 16-50, 190-550, $900-2,000$ coverage 16-50, $190-500,900-2,048$
metres.
Valve lice $\mathrm{X} 79,6 \mathrm{AA}$, metres, Gzsio and two 6AOS. This chassis is suitable for use on A.C. mains from $100-110$ F. and $200-250$ ₹. Price 17 guiness or EA deposit
carriage and insurance $7 / 6$.


## ANOTHER CLEVELAND CHASSIS-"THE TREMENDO"

 The Cleveland Organtone Is good, but this one is really superb. It has a 7 -valve circuit with 6 watts output, fitted with !ndependent bass and treble controls.It is really an efficient R.F. circuit coupled to a high-fldelity amplifier The chassis size is the same as the Organtone, namely $12 \times 7 \times 7$ with the $101 \times 41$ multi-coloured scale, sand it is bulit to the same exacting specification as the Organtone. Price $£ 15 / 10 /-$, carrlage and packjig $7 / 6$. H.P. terms if required.
Ditto but with ferrite rod aerial coil. $£ 16 / 10 /-$.

## THE ARMSTRONG F.C. 48

Among high class radio chassis, the name Armstrong is probably the
moat famous, and their new model RC48 certainly lives up to tradition.
It is virtually a 10 -valve circuit, It ts virtually a 10 -valve circult, for among its eight valves two double
triodes are employed. triodes are employed. Special
features of this chasgis are (a) 8 watts output in a push-pull circutt ensure the highest fidelity: (b) provision for using F.M., e.g., power brought out to sockets and indicator on dial: (c) independent bass and treble controls with visual indication
 and $1,000-2,000$ metres. The size of this chassis is $12 t \times 9 \times 9$ in. Price e83/18/-, plus $7 / 6$ carriage and Insurance.
OFFERED FOR PRICE OF ITS COIL PACK ONLY

This set, a product of one of our famous manufacturers, has
H.F. stage, tuning H.F. stage, tuning
indicator, and all modernrefinements, covers 5 wavebands includlug short Waves to 11 metres. Offered leas valve:, power-pack, scule
and drive, otherwise complete and unused, price E 5 . Or 20/- deposit. balance
months,
over 12
carriage $7 / 6$ (uses octa' range valves).

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Constructed to VHF standards throughout. Covers the band 2 with RF, Mixer, 2-IF, and ratio detector stages. Provision is made for single or push pull output, or added Short Waveband. Although "hand built" in small quantities, an attractive price is maintained.
Model " A," FM tuner. A popular and small unit, with good sensitivity. These are in use from Bognor to Ely, and little changed since first described by Amos and Johnstone in the "Wireless World." New "hammer " finish front plate and tuning seale carries a magic eye; this and power unit are optional.
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Complete with two detachable diamond heads and transformer. Cash Price $20 / 19 / 9$ or sent for 43 Deposit and 10 monthly payments of $40 \%$. Poss and packing paid.
Delivery of all the above is from stock. We can also supply Wharfedale, Goodmans and Tannoy loudspeakers, etc., Connoisseur Variable 3-speed Motors and all other Quality Equipment on EASY TERMS.
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$\star$ Completely separate recording and replay chalns with direct/replay monitor comparison switching.

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The latest PFA unit is built especially for use with our range of Williamson Ampliflers. Separate base and treble controi in equaliser section. Low noise-high gains 5 mv . input, 6 valves.

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$\mathrm{HS} / \mathrm{CR} 3 / 2(800 \mathrm{c} / \mathrm{s}$ and $5,000 \mathrm{c} / \mathrm{s})-\mathrm{E8.10.0}$
$400 / \mathrm{CR} 3 / 2(400 \mathrm{c} / \mathrm{s}$ and $5,000 \mathrm{c} / \mathrm{s})=£ 12.10 .0$
WHARFEDALE half-section Crossover Units are now fitted with a 0.9 mH inductance in series with the middle speaker ( $10-$ 15 ohms), with a switch to short-circuit the coil according to requirements.
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$1,200 \mathrm{ft}$. REEL OF SCOTCH BOYMAGNETICRECORDING TAPE.

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## THE "MINI TWO-THREE"

An " Alldry " Battery Portable o midget size, $61 \mathrm{in} \times 4 \mathrm{in} \times 31 \mathrm{in}$. band 190-559 metres, with use of dhort trailer aerial.
The simple design of this Receive is no arranged that either a 3 -valv aet or a 2 -valve \{afterwards easily made. Consists of a T.R.F. circuit using stage and a bigh gain output pentode. Valve llne up IT4-1T4-DL94 The 2-valve set can be completely built for $84 / 3 / 6$ (less case) and Wach prlce includes valves, speaker and drilled charsis.


A BULK PURCHASE ENABLES THIS SPECIAL PRICE REDUCTION OF THE FAMOUS

## SHAFTESBURY PORTABLE AMPLIFIER



Suitable for home use and small Halls. Has matohed inputs for both Record Players and Mirarophone. Also pro-
vides for the " mixing "and "Iadiag" of both Gram. and speech is request.


## COMPRISING

(a) A 4-Falve High Gain Amplifter for use on A.C. or D.C. maning. $200 \cdot 250$ volts with watts output. Incorporating independent
Volume Controls for Mike and Gram., either of which can be faded at will, a variable Tone Control and independent iaput sockets for Mike and Gram.
(b) A Transverse Carben microphone which obtains its polarizios current from the ampllies -no batteries are necessary.
c) A 8in. Goodmans P.M. Speaker with the "Ticonal" alagnet for

TME COMPLETE EQUIPMENT is all contained in the
portable carrying case $£ 18^{\prime} 0^{\prime} 0$
Having been reduced trom $£ 30 / 9 /$. HIRE PURCHASE TERM DEPOAIT $84 / 10 /$ sad 12 monthly payments of $\mathrm{fl} / 5 / 4$ Light illustrated leaffer containing free data is available on receipt of $\mathrm{S}_{\text {o A. }} \mathrm{B}$.

## 109 wa 115 FLEET ST.

LONDOM, E.C.4. Phone: CENTRALSER2.3.4

- Receiver, requiring H.T above, or approx. to 69 volts. The Kit Is is housed in a light-sluminium case size $41 \mathrm{in} . \times 11 \mathrm{ln} \times 3$ ain. Price of complete kit with easr-to-follow is housed in a inght-suminil. assembly instructions, $42 / 6$. at 250 mA . Size of assembled unit $71 \mathrm{n} . \times 2 \mathrm{fm} . \times 1 / \mathrm{ta}$. Price 47 ib

A DUAL-CHANNEL PRE-AMPLIFIER and TONE CONTROL UNIT
Attractively findshed to " Old Gold" and providinz with a main volume control
It can be used with any amplifier and with any pick up, the range of frequency control provided by the unit affording ainple compenkation for all types o pick-ups and all natures of reconus. The extreme flexibility of the bass and treble control is such that the level of bass and treble can be set to suit any conditlons irrespective of the volume output of the amplifer. Response characteristacs are given in
$12-w a t t$ namplifer advt. The unit measnres only. $9 \mathrm{in} . \times 4 \mathrm{in} . \times 2 \operatorname{lin}$.a including self-contained power supply and can be accommodnted either on or away from the main amplifier on the tront panel of a cabset or any other porice including drilled chassig valves (6sN7, and 6J). E3/18/9. Complete asserably data are available separately for $1 /-$. Completely assembled and
T THE ID A 5-VALVE 2-WAVEBAND SUPERHET RECEIVER OPERATED FROM A
6-VOLT BATTERY FOR ONLY £6'17'6

These Recelvers, which we bave recently acquired by bulk purchase, are ex-Britiah Ministry of Supply, and are new and unused. They are a two-waveband Superhst with R.F. Stege, covering short Wave 18 to 50 metre and Mediam wave 200 int. loudspeatier ls built in and the whole Chasis is contained in a metal cabinet with lid and carrying handle which measures $12 \mathrm{fin} . \times 7 \mathrm{ln}, x$ 71 in . overall. Valve lime up is 7A7, 7Q7, 7A7, 7B6 and 7C5. They possess excellent sensitivity and will give very good resulte on a very short acriai. They are made for to also enable their use from a 12 -volt battery.

WILLIAMSON AMPLIFIERS BY GOODSELL
These Amplifiers hardiy need enlarging upon, belng sufficient to bay that they have now becone tae accepted standard or quality reprodncton by which all others ar udged. Two Models are available: MODEL G.W. 12 Uses slightly lower H.T. voltage to produce 10-12 watts output Price $227 / 10 /=$ (Plus 7/6 Carriage H.P. Terms Deposlt £6/17/6 and 12 monthe at e1118/8 THE MODEL P.F.A. TONE CONTROL UNIT
This Control Unit has established a reputation for ts excellent quatity of reproduction 4. Price $220 / \mathrm{F} /$ - (Plus $7 / 6$ Carriage of pick-up. H .P. Terms. Deposit 55 $220 /-\mathrm{mand}$ Insurance.)
SENS S.A.E. FOR ILLUSTRATED LEAFLETS

## SELENIUM RECTIFIERS

L.t. Types

2/6v. $\frac{1}{\text { a ah.w.... } 1 / 9}$
6/12 v. $\frac{1}{2}$ a.h.w. 2/9
F.W. Bridge Types

6/12 v. 1 a. ...... 5/9
0/12 v. 2 a.
8/9
H.T. Typz H.W. 120 V .40 mA . 250 v .50 mnA 250 v. 150 m RM4 250 v. 250 ma 300 v. 275 mA

CO-AXIAL CABLE. 75 ohms tin., 7d. yard Twin screened feeder, 10 d .
yard.

SILVER MICA EONDENSERS. $5,10,15,20,25$, $30,35,50,100,120,150,180,200,230,300,330$, $400,470,500,1,000$ pid. (.001 $\mu \mathrm{F})$, 002 mfd


DIAL BULBS, M.E.S., \& v. 0.15 a, , $6 / 9 \mathrm{doz}$. 6.5 v. 0.3 a., $6 / 9$ doz.; 4 v. $0.3 \mathrm{a} ., \mathrm{6} / \mathrm{-} \mathrm{doz}$.

ELECTROLYTIOS (current production).

| Tubular Typss |  | Can Types |  |
| :---: | :---: | :---: | :---: |
| F 450 v | 1/9 |  |  |
| 8 mfd . 500 | 2/6 | 16 mfd .350 v .... | 1/11 |
| $16 \mu \mathrm{~F} 350 \mathrm{v}$ | 2/3 | $16 \mu \mathrm{~F} 450 \mathrm{v}$. | $2 / 9$ |
| $16 \mu \mathrm{~F} 450 \mathrm{v}$ | 2/9 | $24 \mu \mathrm{~F} 350 \mathrm{v}$ | 2/11 |
| $16 \mu \mathrm{~F} 500$ v | 3/9 | $32 \mu \mathrm{~F} 350 \mathrm{v}$ | 2/11 |
| $32 \mu \mathrm{~F} 350$ v. | 3/9 | 32 infd. 450 v | /9 |
| 32 mid . 500 v . | 5/9 | (is mind. 450 v | 4/9 |
| ${ }^{8-16 \mu} \mathrm{~F}^{500} \mathrm{v}$. | 4/11 |  |  |
| $2 \mathrm{~L} \mu \mathrm{~F}$ <br> 50 F <br> 12 v <br> 12 | 1/3 | 100 mik | 4/9 $3 / 6$ |
| $50 \mu \mathrm{~F} 50$ | 2/3 | 8.8 mfd .500 v . | $4 / 9$ |
| 100 mid . 12 v . | 1/9 | 8-16رFF 450 v .. | 2/11 |
| 100 mid . 25 v . ${ }^{\text {an }}$ | 2/3 | 16-16 $\mu \mathrm{F} 450$ | 4/11 |
| Gan Types |  | 16-32 $\mu \mathrm{F} 350$ | 4/9 |
| unfd. 450 | 2/3 | $32-32 \mu \mathrm{~F} 350$ | 4/9 |
| \% mf | 3/9 | $32-32 \mu \mathrm{~F} 450$ | 5/1 |

Many others in stock.

## R.S.C. TRANSFORMERS

 FULIY GUARANTEED, INTERLEAVED AND IMPREGNATEDMAINS TRANSFORMERS

## Primaries $300-230-250$ v. $50 \mathrm{c} / \mathrm{s}$.

FULLY SHROUDED UPRIGHT MOUNTING
$250-0-250 \mathrm{v} .60 \mathrm{~mA} .6 .3 \mathrm{v} .2 \mathrm{a}, 5 \mathrm{v} .2 \mathrm{a}$
Midget type, $2 \frac{1}{2}-3-3 \mathrm{in}$.
17/6
$350-0-350 \mathrm{v} .70 \mathrm{~mA} ., 6.3 \mathrm{v} .2 \mathrm{a} ., 5 \mathrm{v} .2 \mathrm{a} . . .119 / 9$ $250-0-250$ v. $100 \mathrm{~mA} ., 6.3$ v. -4 v. -4 a., c.t. 0-4-5 v. 3 a.
$250-0-250$ v. $100 \mathrm{mi} ., 6.3$ v. 4 a., 5 v. 3 a...
$250-0.250$ v. $100 \mathrm{~mA}, 6.3$ v. 6 a., 5 v. 3 a.$~$
for R1355 conversion
$300-0-300$ v. $100 \mathrm{~mA} ., 6.3$ v. 4 a.. 5 v. 3 a... $300-0-300$ v. $100 \mathrm{raA} ., 6.3$ v.- 4 v. 4 a., c.t.
$0-4-5$ v. 3 a.
$31 /-$
$23 / 9$
27/9
$350-0-350$ v $100 \mathrm{~m} .4 ., 6.3 \mathrm{v}, 4 \mathrm{a}_{0}, 5 \mathrm{v} .3$ a...
$0-4-5$ v. 3 a
23/9
$350-0-850$ ч 150 mA . 3 v 4 ........................................
$350-0-350$ v. $150 \mathrm{~mA} ., 6.3$ v. 4 a., 5 v. 3 a...
$350-0-350$ v. $150 \mathrm{~mA} ., 6.3$ v. 2 a., 6.3 v. 2 a 5 v. 3 a.
55 $33 / 9$
4 a., c.t., 5 v, 3 a, suitable w., c.t., 6.3 v
Amplifier, etc. 3 a., suitable Williamson
$450-0-450$ v. $250 \mathrm{~mA} ., 6.3$ v 6 a., 6.3 v...... $49 / 9$
TOP SHROUDED DROP THROUGH TYPE
$250-0$-250 v. $70 \mathrm{~mA}, 6.3$ v. 2.5 a. $13 / 9$ $260-0-260$ v. $70 \mathrm{maA}, 6.3 \mathrm{v}$. 2 a., 5 v. 2 a.... $16 / 9$ $350-0-350$ v. $80 \mathrm{~mA} ., 6.3$ v. 2 a., 5 v. 2 a.... $18 / 9$ $250-0-250 \mathrm{v} .100 \mathrm{~mA} ., 6.3$ v. 4 a., 5 v. 3 a.... $22 / 9$ $300-0-300$ v. 100 mA., 6.3 v.-4 v. 4 a., c.t., $23 / 9$
$0-4-5$ v. 3 a. ....................................... $23 / 9$ 0-4-5 v. 3 a.
3 a
3
350
$350-0-350$ v. 1
$0.4-5$ v.
$350-0-350 \mathrm{v}$
5. v. :3 a

350-0-350 v. 150 IuA., 6.3 v, 4 a., 5 v. 3 a.... $29 / 9$
E. H.T. TRANSFORMERS, 2,500 v. 5 .nA.
, ${ }^{2}$ RANSFORMERS, 2,500 v. 5 MA.
2-0-2 v. 1.1 a., $2-0-2$ v. 1.1 a., for VCR97,
VCR517

FILAMENT TRANSFORMERS
Primaries $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$.

| 6.3 v. 1.6 a | 5/9 | 0-4-6.3 v.2 а.... | $7 / 9$ |
| :---: | :---: | :---: | :---: |
| 6.3 v, 3 a , | 8/11 | 0-2-4-5-6.3v. 4 a | 16/9 |
| 12 v .1 | 7 | 6.3 v. 6 a. | 17/6 |
| 6.3 v. 2 a. | 7/6 | 1.5 a. ......... | 17/6 |

## CHARGER TRANSFORMERS

All with $200-230-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. Primaries: $0-0-15 \mathrm{v}$ 14 a., $11 / 9 ; 0-9-15$ v. 3 a., $0-3.5-9-17$ v. 4 a., 18/9. 0-9-15 v. 5 a., $19 / 9 ; 0-9-15$ v. 6 a., $23 / 9$

## ELIMINATOR TRANSFORTHERS

Primaries $200-250$ v. $50 \mathrm{c} / \mathrm{s} .120 \mathrm{v}, 40 \mathrm{~mA}$. 130 v. $50 \mathrm{~mA} ., 6.3 \mathrm{v} .3 \mathrm{a} .$.
120 v. $40 \mathrm{mA},. 5-0-5 \mathrm{v} .1 \mathrm{a}$
90 v .15 mA, , 6-0-6 v., 250 m
7/11

## OUTPUT TRANSFORAERS

Midget Battery Pentode $06: 1$ for 354 , etc
Small Pentode, $5,000 \Omega$ to $3 \Omega$
Standard Pentode, $5,000 \Omega$ to $3 \Omega$
Standard Pentode, 8,000 to $3 \Omega$
Battery Pentode, 10,000 ohms to 3 ohms.
Multi-ratio $40 \mathrm{~mA} .30: 1,45: 1,60: 1,90: 1$
Class B Push-Pull
Push-Pull 8 Watts 6 V 6 to 3 ohms
Push-Pull 10-12 Watts 6 V 6 to $3 \Omega$ to $15 \Omega$,
sectionally wound .........................
Push-Pull $10-12$ Watts to match ove to
3-5-8 or 15S
Push-Pull $15-18$ Watt, sectionally wound,
6L6, KT66, etc., to 3 or 15 ohms..
Push-Pull 20 Watt high-quality sectionally
wound, $6 \mathrm{~L} 6, \mathrm{KT} 66$, etc., to 3 or $15 \Omega$.
SMOOTHING CHOKES
$250 \mathrm{~mA} .3 \mathrm{H}, 100$ ohms
$250 \mathrm{~mA}, 3 \mathrm{H}, 100$ ohms ...
$150 \mathrm{~mA} .$,
$100 \mathrm{~mA} ., 10 \mathrm{H}$.
$10 \mathrm{H} ., 150$ ohms
150
$100 \mathrm{~mA} ., 10 \mathrm{H} ., 150$ ohms potted.
$100 \mathrm{~mA}, 10 \mathrm{H}, 200$ ohms
$30 \mathrm{~mA} ., 10 \mathrm{H}_{.}, 350$ ohms
$60 \mathrm{mA},. 10 \mathrm{H}_{-,} 400$ ohms

VOLUPE CONTROLS with long spindles, all values, less switch, $2 / 3$; with S.P switch, $3 / 9$.
WIRE WOUND PDTS: 30 ohms, 500 ohms, $5 \mathrm{~K}, 20 \mathrm{~K}, 100 \mathrm{~K}$ (medium length spindles), 2/9. 290 ohms, $91 \mathrm{~K}, 10 \mathrm{~K}$, 20K, Preset type, $1 / 9$ each.
VIBRATORS. Wearite 12 v. 4 pin. Non synchronous, 6/9. Dak 2 v. 7 pin, syn chronous $7 / 9$

EX GOVT. E.H.T. SWOOTHING CONDENSERS .25 mifd. $4,000 \mathrm{v}$. Blocks
$4 / 9$
$3 / 9$
.5 mfd , $2,000 \mathrm{v}$ Blocks
$5 \mathrm{mfd} ., 3,500 \mathrm{v}$. Cuns
1 mfd . plus 1 uifd. 8,000 v., large blccks
(common negative isolated)
3/3
$1.5 \mathrm{mfd} .4,000 \mathrm{v}$ Bloch
EX GOVT. METAL BLOCK PAPER CONDENSERS
2 mfd .800 v .... $1 / 9 \quad \mathrm{f}-6 \mathrm{mid} .450 \mathrm{v} .5 / 9$

4 mid. 500 v.... $2 / 3 \quad 8$ nifl. 500 v
4 mid. 1,000 v. $\quad 4 / 3 \quad 8.8$ míd. 500 v
4 mifd. 400 v . plus 2 mid . 250
EX. GOV. ISNITS, type RF26 in original sealed cartons 39/6. Transmitter Receivers type TR9D complete with all valves $45 /-$, carr. $6 / 6$.
M.E. SPEAKERS. All 2-3 ohms, 8 in. R.A. field' (100) ohins, $11 / 3$. 1 Cin. K. 4. field. 1,500 ohms, $23 / 9$. 10 in . R.A. field, 1,000 ohms, $23 / 3$.
MANUFACTURERS SURPLUS
TRANSFORMERS
Fully shrouded upright. Primaty 200-230.250 v Fully shrouded upright. Primaty $200-230 \cdot 250 \mathrm{v}$.
Sec. $.425-6-425 \mathrm{v} .150 \mathrm{~m} . \mathrm{a} \cdot 6.3 \mathrm{v} .3 \mathrm{a} .6 \mathrm{v} .3 \mathrm{a} .37 / \mathrm{s}$.
GOODMANS $3 \frac{1}{2} \mathrm{in}$. P.M. SPEAKER (ex, equip.), with battery pentode trans., $12 / 9$.
HEAVY DUTY BATTERY CHARGER
For normal $200 / 250 \mathrm{v}$. A.C. mains input. To charge $12 v$. battery. Varlable charge rate of up to 10 amps . Fitted Meter and Fuses. Guaranteed 12 Inonths. Care. 10/-. 86/19/6.
SOIL FILLED BLOCK CONDENSERS
Bryce $11-7 \mathrm{mfd} .500 \mathrm{v}$. New unused Govt. surplus, only $5 / 9$ each.
H.T. ELIMINATOR AND TRICKLE CHARGER KIT with louvred crackle finished case. Mains imput 200-250 v. Output $120 \mathrm{v}, 40 \mathrm{~mA}$., and 2 v . $\frac{2}{2}$ a. Price with circuit, $29 / 6$.
Or in working order, 37/6.

## THE SKY FOUR T.R.F. RECEIVER



## EX GOVT. MAINS TRANSFORMERS

All $230 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. input.
8.8 v. 4 a.

88 v. 1 a. ...................
$300-0-300$
$278-0-278$
v.
80
80 mLA.
9/9
Cartiage on following types 51 $0-11-22 v, 30$ a. ........................ $72 / 6$ $16-18-20 \mathrm{v} .35 \mathrm{a}$. 7.7 v. C.T. 7 amps., 4 times. 400 v .200 mA .6 .3 v .5 a .
$300-0-300$ v. 150 mA ., $610-0.610$ v. 150 mA $1,200 \mathrm{v}, 250 \mathrm{~mA}$.
400 v. C.T. 150 mA .4 v. 5 a., 6.3 v. 6 a.
6.3 v. $0-6$ a., 4 v. 6 a., 4 v. ba., 4 v. 3 a.,

4 v. 3 a., 5 v. 2 a.
$22 / 9$
$29 / 9$

| $325-0-325$ | v. $150 \mathrm{~mA}, 6.3$ v. $4-6$ a., 5 v. 2-3 a. | $22 / 9$ |
| :--- | :--- | :--- | :--- |

## EX GOVT. AUTO TRANSFORMERS

15-10-5 -0-105-215-235 v. 500 watts.
Double wound $10-0-200-240 \mathrm{v}$. to $10-10-275$. $205-315$ v. 1,000 watts
Double wound $0-110-240$ v to $0.130-140$
$150-160-170$ v. 1,500 watts
69/6

Carriage on any of above $5 /$-extra
EX GOVT. SMOOTHING CHOKES
$250 \mathrm{~mA} ., 10 \mathrm{H} ., 50$ ohmas
$14 / 9$
250 mA ., 10 H .100 ohins $14 / 9$
$8 / 9$
250 mA ., $3 \mathrm{H} ., 50$ ohins
150 mA ., $10 \mathrm{H}, 50$ ohms
100 mA ., 10 H ., 100 ohms, Tropicalised
$100 \mathrm{~mA} 5 \mathrm{H}, 100$ ohms, Tropicalised
50 mA ., $50 \mathrm{H}, 1,000$ ohnis, Potted
$90 / 100 \mathrm{~mA}$., $10 \mathrm{H} ., 100$ ohms, Potted
$50 \mathrm{~mA} .5-10 \mathrm{H}$.
$50 \mathrm{ma}, 5-10 \mathrm{~h}$.
L. T. type 1 mp

A design of a 3 -valve $200-250$ v. A.C. Mains receiver with selenium rectifier. For inclusion in either of cabinets illustrated above. It employs for simplics, SP61, 6F6G, and is specially desioly is sell up to standard Sensint-to-point wiring diagrams instructions and parts list, 2/3 This receiver can be built for a maximum of $£ 4 / 19 / 6$ including cabinet. Available in brown or cream bakelite, or veneered walnut.
P.M. SPEAKERS. All $2-3$ ohms. 61 in . Plessey 8 in 5,000 ohm output transormer, Rola with Trans., 29/6.
R.S.C. BATTERY CHARGER KITS. For mains input $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. To charge 6 v . accumulator at 2 amps., $25 / 9$. To charge $6 v$, or $12 v$. battery at 2 a., $31 / 6$. To charge 6 v. or 12 V . battery at ${ }^{4}$ a., $49 / 9$. OF GREEN CRACKLE LOUVRED STEEL CASE, MAINS TRANSFORMER, FULL WAVE METAL RECTIFIER, FUSES, FUSE-HOLDERS AND CIRCUIT. Any type assembled and tested for $6 / 9$ extra.
R.S.C. $6 v_{0}$ or $12 v_{0}$ BATTERY CHARGER For rormal A.C. mains input 200-230-250 v., 50 $\mathrm{c} / \mathrm{s}$. Selector panel for 6 v. or 12 v. charging. Variable charge rate of up to 4 AMPS. Fused, and with ammeter. Well ventilated metal case with attractive crackle finish. Guaranteed for 12 months, 69/6, Carr. 2/6.


## CHASSIS

18 s.w.g. undrilled aluminium amplifier type ( 4 -sided).
$14 \mathrm{in} \times 10 \mathrm{in} \times 3 \mathrm{in} .7 / 11$ $16 \mathrm{in} . \times 10 \mathrm{in} . \times 3 \mathrm{in} .8 / 3$ 18 s.w.g. aluminiun recelver type.
6 in. $\times 3$ in. $\times 1 \frac{\mathrm{in} .}{} 1 / 11$ 7 in $\mathrm{in} . \times 4$ tin. $\times 2$ in. $2 / 9$ $10 \mathrm{in} \times 5 \overline{2} \mathrm{in}, \times 2 \mathrm{in} .3 / 3$ $11 \mathrm{in} . \times 6 \mathrm{in}, \times 2 \frac{1 \mathrm{in}}{}$. $3 / 11$

18 s.w.g. aluminium receiver type.
$12 \mathrm{in} \times 8 \mathrm{in} . \times \frac{21}{2} \mathrm{in} .5 / 3$ $16 \mathrm{in} . \times 8 \mathrm{in} . \times 2!\mathrm{in} .7 / 6$ $20 \mathrm{in} . \times 8 \mathrm{in}$. $\times 2 \frac{1}{2} \mathrm{in} .8 / 11$ 16 s.w.g. aluminium anmplifier type, 4 -sided.
$12 \mathrm{in} . \times 8 \mathrm{in} . \times 2 \frac{1}{2} \mathrm{in}, 7 / 11$ $16 \mathrm{in}, \times 8 \mathrm{in} . \times 21 \mathrm{in} .10 / 11$
$20 \mathrm{in} . \times 8 \mathrm{in} . \times 24 \mathrm{in} .13 / 6$

## R.S.C. HIGH FIDELITY 25 watt AMPLIFIER A4

## A NEW DESIGN FOR 1955

 HIGH GAIN "PUSH PULL OUTPUT". BUILT-IN PRE-AMP. TONE CONTROL STAGES. INCLUDES 7 valves, sectionally wound output transformer, block paper reservoir condenser, and reliable small components. AN INPUT OF ONLY 20 millivolts IS REQUIRED FOR FULL OUTPUT. THIS MEANS THAT ANY TYPE OF MICROPHONE OR PICK-UP IS SUITABLE. Two separate inputs controlled by separate volume controls allow simultaneous use of "Mike" and Gram. or Tape and Radio, etc., etc. Individual controls for Bass and Treble "lift" and "cut". Six negative feedback loops giving total of 24 D.B. Frequency response $\pm 3 \mathrm{D} . \mathrm{B} .30-20,000 \mathrm{c} / \mathrm{s}$.

Hum level 66 D.B. down. Certified total harmonic distortion of only $0.35 \%$ measured at 10 watts. Comparable with the very best designs. SUITABLE FOR SMALL HOMES OR LARGE HALLS, CLUBS, GARDEN PARTIES, DANCE HALLS, etc., etc. For ELECTRONIC ORGAN OR GUITAR. For STANDARD OR LONG PLAYING RECORDS. Size $12 \times 10 \times 9$ in. For mains A.C. $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. Power consumption 175 watts. Outputs for 3 and 15 ohm speakers. The kit is complete in every detail. Chassis is fully punched. Easy to follow point-to-point wiring diagrams are supplied. EXTRA HIGH SENSITIVITY, HIGHEST QUALITY for 9 GNS .
Or assembled ready for use $50 / \mathrm{e}$ extra. 9 GN
H.P. Terms on assembled units. Deposit $26 /=$ and 12 monthly payments of £1. Plus carr. ro/-

Terms to include cover, mike, speakers, etc., on request. Cover as illustrated if required, price $17 / 6$ extra

## A PUSH PULL 3-4 WATT HIGH GAIN <br> ASSEMBLED AMPLIFIER FOR $£ 3 / 19 / 6$.

 For mains input $200-2500$ r. 50 cts . Complece kit of pastsincluding polint-to-point wiring diagrams and insuruections. including point-to-point wirink diagranns and insuructions.
Amplifer can be used with any type of feeder unit or pickAmplifier can be used with any type of feeder unit or pick-
up. This is not A.C./D.C. with " "live " chassis but A.C. up. This is not A.C./D.C. with "live" chassls but A.C.
only with $400-0-400$ v. Trins. Output is for $2-3$ olim speaker. Supplled ready for use, $\mathbb{Z} / 19 / 6$. Fulk descriptive leadlet 6 d .
H,M.V. LONG PLAYING RECORD TURNTABLE COMPLETE WITH CRYSTAL FICK-UP (SAPPHIRE STYLUS;. Speed 33i r.p.m. BRAND NEW, CARTONED. Ouly £3/19/6 (approx. hal \& price). Carr.
R.S.C. A7 3-4 WATT QUALITY AMPLIFIER A highly sensitive 4 -valve amplifer using negative feedback and having an exem lent frequencyresponse. Pre-ampliner and Tone Control stages are Incorporated with separate for Long Playing records. Sultable for many kind of plek-up including latect high fidelity types. H.T. of 2500.20 mA . aud L.T. 6.3 v. la, available for supply of Radio Feeder Unit, etc. ONLY 40 millivolts input required for full output. Fully isolated charsls with baseplate. For A.C. maling $200-$ wiring diagrams and Instructions. Only $£ 3 / \mathbf{1 5} / \%$
R.S.C. 4-5 WATT HIGH GAIN AMPLIFIER


A highly sensicive 4 -valve quality amplifier for the home. sinall elub, etc. Only 50 milluvoits input is required for full output so thaty plekp heads. In zeddition to all othen types of plek-ups and procticsly all mikes, Separate Bass and
Treble controls are provided. These give fuli long playing Treble controls are provided. These give foli long playing
record equalistion. Hum level is neglisible being $71 \mathrm{D.B}$. record equalis;tion. Hum level is negligible being 71 D.B.
down. $15 \mathrm{D} . \mathrm{B}$. of negative teedback in usel. H. of $300 \vee .25 \mathrm{~mA}$. and L.T. of 6.3 ved. 1.5 k . Is available for the supply of a Rulio Peeder Unit, or Tape Deck pre-amplifier. For A.C. msins input of $200 \cdot 230-250 \mathrm{~V}$. 50 o/s. Cunse fully punched chassls (with baseplate), with green crackle finish, and poinl-to-polint wiring diagrang and instruetions, Exceptional value at only \&4/15/-, or assembled ready for use $25 /$ - extra, plus $3 / 6$ carr. Output for 3 -ohm spesker.

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 High impedance type. Liznited number, brand new, boxed and perfect at iraction, of normal price. Only $35 /-$DEFIANT RECORD PLAYING TURNTABLE COMPLETE WITH MAGNETIC PICK-UP. Pick-up is high impedance type. Onit is housed in a beautiful wainut veneered cabinet of attrantive design. For al standard records ( 78 r.p.m.). Limited number. Brand new, cartoned, $£ 5 / 19$. Carr. 7/6. ACOS HIGH FIDELITY CRYSTAL MICRO
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All parts for an " All Dry" Mattery Eliminator, Complete with case. Completely repiaces 1.4 V , und $90 \quad$ v. batteries wherc normal
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Price with citcouit, 38/9.
 Suitable for recelvers with L.T. loado of 125 mA . to 250 mA . thereby covering latest low ponsumption ty pes.
BATTERY SET CONVERTER KIT. All parts for oon. vertlog any type of battery receiver to all wains. A.O. 2000 $250 \mathrm{\nabla} .50 \mathrm{c} / \mathrm{s}$. Kit will supply fully smoothed E.T. of $120 \mathrm{\nabla}$ 90 . or $60 \nabla$. at up to 40 mA , and fully smoothed L.T. of 2 v. it 0.4 . to 1 . Price complete with circuit and instruc
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Large gafety fintors in every component A.C. and H,T pluge, 6 valves, and with easy-to-follcw point-to-point wiring diagrams. Everything supplied w lost nut. Two independent lnputa are provided with two assoclsted odependent volume controls so that progranmes can be moxed together if desired, sucia as microphone announce-
ments superimposed on is musical programane, or two ments superimposed on in musical prograntne, or two phope/radio, fadiag over from one to the other. Viritible base lift and out with variable treble lift and cut tone goutrols are fitted, giving fill long playing reccril equalis. hion for uncorrected plek-ups. They are also provided so bat the user can alter the tonsl value to sult his personal asto and surroundings. Terminals aro providerl for 3 hm and 15 ohm loudspeakers.
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Six Negative Feedback Loops.
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 Design of a HIGH FDDEfITY L. and M. wave T.R. Unit with geif-contained heater supply and horcugh ir. fron main anapliper. Three valves and Low Dlaturtion Cermanium Diods Detentor. Flat topped response characteristio. Loaled H.F. zoils. Two variable Mu controlled E.F. atages, 3 gans condenser tuning. Cathode fullower oulput stasge. Siritch positlon for Gram. and Gram. input and outpul sockets. Derforminnce comparable with the best in Fecler Unils. For A.C. mains 200-230-250 v.opemtion. Slzo $1[-6.71 \mathrm{la}$. Ilustration, full set of easy-to follow wiring diagrams and instructions and individually priced partalist 2/6. This unll can be built for only $29 / 15 /-$ inoluding Dial and Drise Knobs and every item required.
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Plua $10 \%$ paeking and carriake. Pluan $10 /$ - packing and carriake.
R1124 RECEIVER UFT. COrerage $30-40$ Mc/e. lachuding 6 valves-3 type 9D2, 1 each, 8 D , 16 DD 2 and 4 D 1 -Six valye screening cans, 24 ceramic trimmers, 6 cerannic valye holdera, reeistors, condensers I.F.T.'s colls. ete. In very good sondition.
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$350.0 \cdot 350 \mathrm{I} 30 \mathrm{~mA}$. Tapped $350 \cdot 0 \cdot 350130 \mathrm{~mA}$. Tapped filsunent winding 6 v. 3 A., 15 \%. 3 A., 21.5 ₹. . 6 A., abso | $5 \nabla .2$ A. Troplcallsed drop-through type. |
| :--- |
| $21 /-$ plus $2 / 6 \mathrm{P}, ~ \& ~$ | 6.3 V. 3 a., 5 V. 2 a., 200/250 \%. Input

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Wo are plea sed to announce our complete Kit lor the "Denoo" F.M. Feeder Unit. feeding into the audio section of a standard broadcast receiver where triode/pentode output are available. Within 8 n average of 30 miles from \& V.E.F. transmitter one I.F. Eise should be adequate, but our complete valves for an extra I.F. utage if necessary, or If the unit is msed at grester distances Full Constructional details, theoretical oirouit and point-to-polint wiring diakram can bo
supplied for $1 / 6$ post free or the complete supplied for $1 / 6$ post free. or the complete Kit rixht down to the last nut and bolt at assembled, aligned and tested, at e8/10/-
 plus 2/6 packing and postare.
If required we shall be pleased to align this unit for constructors not possessing the necessary Chuipment for a chatge of $7 / 6$, N.B_-Valve line-up is 8AM6, 12AH8, 2-6BA6 and 6AL5. Chassis measure onstrations at 18 ,
Demonatrations at 18, Tottenham Cour $\mathfrak{k}$ Roadll

## The Jason r.M. Tuner Kit!



This kit has been based on the booklet by Data Publications, price 2/- post free. With each booklet is enclosed our individually priced parts list. The construction and alignment of this tuner are no more difficult than a normal medium wave tuner. It is highly sensitive and free from dritt. Incorporates 4 valves type 6 AM6 and 2 specially graded G. B.C. Crystals. The kit supplied includes drilled chassis with tuning condenser, scale calibrated in megacycles, and attractive bronze stove enamelled front plate already mounted (as illustrated) front plate size 8 in. $\times 5$ in., chassis size 7in. $\times 4 \frac{1}{2} \ln . \times 1 \frac{1}{2}$ in
N.B. The standard model is at present operating satisfactorily up to 80 miles from Wrotham. Our price for the complate standard kit is $86 / 15 /$ - only ! Plus $2 / 6 \mathrm{p} . \&$ p. Fringe area model including extra valve, coil etc. (results could be expected up to 150 miles from Wrotham 1 , is $£ \% / 15 /-$, plus $2 / 6 \mathrm{p} . \& \mathrm{p}$. The Standard Model Tuner can be supplied ready built, aligned, tested and manufactured by the Jason Motor and Electronic Company at a price of £15/17/-, purchase tax paid.
N.B. THESE TUNERS ARE BEING DEMONSTRATED AT 18 TOTTENHAM COURT ROAD.
F.M. AERIALS. Indoor two-element type by Jumex. Brand now $11 / 6$ each only, plus F.M. POWER PACK KIT.-We can now aupply complete kit for power pack sultable or either of the above F.M. tuners or any other similar typs. Price for the corupletie kit 842 i- only, or $52 / 6$ for rowdy assembsed unit. This piack is extremely small incorporatigg for power pack, Bulgin Octal Plug at 2/3.

| F.S.D. |
| :---: |
| 50 microamp |
| 100 microamp |
| 500 microamp |
| 500 microamp |
| 1 mA . |
| 1 mA . |
| 1 ma . |
| 5 ma . |
| 10 mA . |
| 10 mA . |
| 50 mA 。 |
| 150 mA . |
| 200 mA . |
| 1 amp . |
| 3 amp. |
| 5 amp. |
| 6 amp. |
| 20 amp . |
| 25 amp . |
| 30 amp. |
| 15 volt |
| 20 valt |
| 15-0-15 volt |
| 150 volt |
|  |

R.P. - Round Projection
M.C. $=$ Moving Coll. Thermo $=$ Therno-coupled
F. Sq. Fush square. P.R. © Elush Round. MI.I. = Moving Iron

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We are pleased to announce advan-
tapeous hire purchase lacilities on tareous hire purchase gacilities on
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V. 4 pin Mallory Vibrators,
transformers, condenaers, resistors, signs transformers, condenners, resistors, signs
1 amp . indicator, etc., etc., in good con 1 amp . indicator, etc., etc., in good con dition. Complete in metid box size 10/in. $x$
bin. $\times 8$ in. Weight 191 b ., 27/6. plus $5 /$ P. \& $P$.
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 6 or 12 v. 1.5 a. F.W. bridge type.. $9 / 8$
6 or 12 v. 2 a. F. W. bridge type .... $11 / 3$
 CHARGER TRANSFORMERS, Input 230 r 6/12\%. 1 a.
$\begin{array}{llll}2 / 6 / 12 & \text { v. } 2 & \text { a. } \\ 2 / 6 / 12 & \text { v. } 4 & 3 .\end{array}$
ACOS TYPE 7 Crystal Mierophone Inserts Brand new, METER SPECLAL': We havel a limited quatity of aircrait electri, al thermometere meter, flush square fitting. These meter have a luminous scale graduated 40.140 degrees centigrade, but the full scable defleo tion is approximately 150 microamps!
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HRADPEONES. Brand IeT, ax-Govth, by B. G. Brow. Type OLR. Low tesigtance. 716 pas palr. Type CEB high resistance, PORTABLE CABINETS.

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surplus. Well made brown rexina covered. Will take any atandard single player with bottom clearance of 3ln. Total glze clased 15in. $\times 13$ inn. $\times$ 5ilin., fitted Fith sonp plua $2 / 6$ 8. and $\mathbf{P}$.
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TEE NEW R.C. HEGB-FIDELITY A PLIFIER. P.P. $6 V 6$ ou cupt. Freq. 25 18,000 eps 60 db at 61 watts. Treble tion. Provision for Fecder Unit Max UNDISTORTED OUTPUT 8 t watta, Price 14 gns. plus 7'6. NOW AVAILABLE

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Carrying cases in black leatherotte fnish. An extremely well-made case with chrome locks and corner-pleces for extra strength,
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We have perhaps the most up-to-date valve
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spindle. dritcc spring and knobs, at $451-$ plua $2 / 6$ pack ing and casciage
N.B. - Our kits are even supplied with sufficient solder for the jab.
N.B. All our T.R.F. Kit circalts now include specially wound Dencofy Maxi-Q" colls on polyst yrene formers, improved performanee ! Price remains the same.

THE "ECONOMY FOUR" T.R.F. KIT A three-valve plus metal rectifer receiver. can supply all required components right do the last nut and bolt. Valve line up $6 \mathrm{KK7} 6 \mathrm{J7}$ and 6V6. Chassis ready dritiled-Cabinet size 191n. Iong by 6in. high by 5in. deep-Chaice of ivory or brown Bakclite, or wooden, walnut finamh cabinet. Complete instruotion booklet with practical and and teated prior to packing. Our price £5/10/-complete-Remenber this set is being demonstrated at our shop premises! We proudly claim thaw our fully Hlustraved instruction bookletis the
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## MAINS TRANSFORMERS

 $300-0=300,100 \mathrm{~mA}, 6 \mathrm{~F} .3$ amp. 5 v. 2 amp., 22/6.Semi-Shrouded, drop-through 380-0380 v., $120 \mathrm{~mA} ., 6.3$ マ. 4 amp., 5 v., 2.5 амрр., $22 / 6$.

Drop thro' $350-0-350$ v. 70 mA . 6 v . $2.5^{\circ} \mathrm{amp}$., 5 v. $2 \mathrm{amp} ., 14 / 6$.
Chassis mounting or drop-thro'. Pri. 110250 v . Sec. $350-0-350,250 \mathrm{~mA}$.,
 $32 / 6$.
Charsis mounted and fully shrouded, $80 \mathrm{~mA} ., 6$ ष. 3 amp., 5 จ. 2 amp.. $14 / 6$. $250-0-250 \quad 80 \mathrm{~mA} ., 6$ จ. $4 \mathrm{amp} ., 14 /$-.
Drop thro' 270-0-270, 80 mA .6 v. $3 \mathrm{amp} ., 4$ v. $1.5 \mathrm{amp} ., 13 / 6$.
Drap thro' 270-0-270 60 mA .
3 amp., 11/6.
250 จ. 350 mA .6 .3 จ. 4 a., twlec 2 v
Anto-trans. Output $200 / 250$ H.T. 500 V 250 mA., 6 v. 4 a., twice, 2 v. 2 a., 19 ' 6.

Auto Trsins. Input 200/250. H.T. 350 v. 350 mA . Soparate L.T. 6.3 v.

Primary, 230 v., tully shrouded, screened primsry, 13 v. $1 \mathrm{amp} ., 7 / 6$.
Prim 200 v- Seo. 500-0- $\overline{0} 00$ and 500-0-500
 v. 3 ,

Mains Transformer, fully impregnated. nput $210,220,230,240$. Sec. $350-0-350$ 100 mA . With separste heater trans6.3 v. 2 sump., 6.3 v. 3 amp., 4 v. 6 armp. $350-0-25075 \mathrm{~mA} .6 .3$ จ. 3 a. $\operatorname{tap} 4 \mathrm{v}$. $50-0$-is
$6.8 \mathrm{v} .1 \mathrm{~A} ., 13 / 6$.
$500-0-500125 \mathrm{~mA} .4$ ष. C.T. 4 ล., 4 จ. C.T. 4 a., 4 v. C.T. 2.6 a., $27 / 6$.
$500-0-500250 \mathrm{~mA} .4$ จ. O.T. 4 a
C.T. 6 a., 4 จ. C.T. 4 a., $39 / 6$.
6ym. M.E. Spenker, 1,000 ohm fiold. 15h.
R. \& A. T.V. enerkised 6łin. speaker 9/6. P. \& P. 2/6.
R. \& A. GHin, M.E. speaker, with O.P. Volume Controls. Long spindles lems volthe $50 \mathrm{~K}, 500 \mathrm{~K}, 1$ meg., $2 / 6$ each.
P. \& P. 3d. each

Volume Control s. Long spindie and switeh, ${ }^{2}, \frac{1}{2}, 1$ and 2 meg., $4 /-$ each.
10 K and $50 \mathrm{~K}, 3 / 6$ ewch. $\frac{1}{2}$ and 1 meg., loug spindle double pole swltch, miniature, 5 - P. P. \& P. зd. ench.
Trimmer: $5-40 \mathrm{pf}$., $5 \mathrm{~d} \quad 10-110,10-250$,
$10-450$ of., 10 d.
Twin-Gang . 0005 Tuning Condenser, 5/With trimmers, $7 / 6$.
Twin Gang, .0005. whth feet, size $3 \times 3 \times 1 / 1 \mathrm{n} ., \quad 6 / 6$
3-gang . 0005 . with feet, size $41 \times 3 \times$ 1fla., $7 / 6$.
T.V. Coils, noulded former, iron-cored wound for re-winding purposes only.
Ali-can $14 \times 1$ in., $1 /$ eaich, 2 imon-core Ali-can $1 \frac{1}{x} \times 1$ inn. $1 / \frac{1}{6}$ each,
Used Meta! Rectitier, 250 v .150 mA .;
$6 / 6$. Metbi Reotifer, $230 \mathrm{v} .45 \mathrm{~mA}, 6 / \%$.
8/6. Mersl Reotifiex, 230 v. 45 man., $6{ }^{\circ}$. meta
$3 / 6$.
OUTPUF TRANSFORMERS, Standard ype 5,000 ohma impo, $49 ; 48.1$ Wish cuara tsed-back windings, $4 / 3$. Minis, 7.000 and $14,000,5 / 8 . \quad 10 \cdot$ wratt prish-
prall, 676 matating, $7 / . \quad 90.13$ ohm pueech ooll, $8 / 6$.
PU8H-BAGK CONNECTING WIRE. Doz. yds., $1 / 6$. Pbat paid.

STANDARD WAVE-CHANGE SWITCEBEs, 4 -pois 3 -way, $1 / 8 ; 5$-pole, 2-way, 19: ${ }^{3}$-pole, 3 -way $1 / 9 ; 9$-pole apinay, 3-yole 4-was, 4 -pole 3 -wing and 4 sinde 3 -pole $4 / 6$ each. 4 -poue 2 -pole 11 -wad twin-watar 5 H: 1 -pole 12 -way single wafer $5 /=$ P. i P. 3 d .


PERMEABLLITY TUNED T.V. UNIT laput 300 ohm bal anced line. coverage $84 \mathrm{Mc} / \mathrm{y}-88 \mathrm{Mc} / \mathrm{s}$ and $174 \mathrm{Me} / \mathrm{m}-217 \mathrm{Mc} / \mathrm{s}$. Vlsion I.F.:-45 Mc/s. 8ound $40.5 \mathrm{Me} / \mathrm{s}$. Uses s3 mizer, and 8CA
Oscilstos, Provision Oscillator, Provision for suto-gain control,
Dimensions gin. wide 6 inn. deep, 4 in , high, 9 in , blank-scale. Width inolnding ecale-overlap Dimensions gin. Wide meablity tuned. Complete with 3 valves. Pout and Pkg. 3/-. \&2/19/6.
T.V. CONVERTER for the now commerciab statlons, complete with 2 valves. Frequency can be set to any channel within the 186-196 Mc/s hand. I.F. Will work into any existing T, V recelver between $42-68 \mathrm{Mc} / \mathrm{s}$. Input arranged for 80 ohm feeder. EF 80 as RF ampliber, ECC81 as local ogcillator and mixer, The gain of the first stage, R.F. amplifier 10DB. Required power supply of $200 \mathrm{D} . \mathrm{C}$, at $25 \mathrm{~mA}, 6.3 \mathrm{v}$. A.C, at 0.6 gmp . Input fliter enfuring ireedom from work into aoy T,H.F. of superhet. Size $4 \frac{1}{2} \times 24 \times 2$ in. P. \& P. $2 / 6$. $£ 2 / 19 / 6$.
R.P. E.H.T. OSCILLATOR COIL. 7 6-9 KV with EYEl rectifer winding, and ol reuit diarram, $15 /{ }^{\circ}$ As above but complete with 676, EY51 and associ ated resistor a and condensers. Circuit diagram 37/6. As a bove but complete with line O.P. transformer, scan coils and freme O.P. tranformer P. 31.

PLASTIO CABINET, as illustrated, 11 in. $\times$ $\underset{\text { polished }}{ } \times \frac{54 \mathrm{hn} ., \text { In Waluut and Cream, blso In }}{\text { Wainut ocmplete with }}$ polished Walnut ecmplete with T.R.F. new wave-hand, back-plate, drum, polnter, apring, drive spindle, 3 knobe and back. 22/6. $P$. \& P. 3/6. AS ABOVE, with superhet chnsiat, 23/6. P. \&P. 3/6. Elther of the above iteins complete with 5 " P.M. speaker and O.P. transformer, $17 / 6$ extra.
Used metal rectitier, $230 \mathrm{v} .50 \mathrm{~mA} ., 3 / 6$;
gang with trimmers, 6/6: M. and L.T.R.P. gang with trimmers, 6/6: M. and L.T.R.F. and circult, $4 / 6$; heater, trans, , $6 /$-: volume control with switch, 3/6: wave-change switoh. $2 /-; 32 \times 32 \mathrm{mfd}$. . $4 /-\mathrm{i}$ blas condenser,
 Used A.C. mains $200 / 250$ volis, 4 valve plus metal rectifier, medium wave superhet in polished
wralnat cabinet, size $14 \times 9$. $\times 71^{*}$, complete with valvea $6 \mathrm{~K} 8,6 \mathrm{~K} 7,6 \mathrm{Q}$ and 6 F 6 . $61 \mathrm{P} . \mathrm{M}$. speaker, Fully guaranteed. P. \& P. 7/6. £3/15/-.
 EXTENSION SPEAKER in polished walnut, complete with 8in. P.M. P. \& P. 3/. 24/6-
SINGLE SPEED PLAYER. A.C. Malns 200/250 V., complete uth needle armature plok-up in a really wonderful pollahed walaut cabinet, will take up to as 12 inch record. Pull-out draw on steel runners. Original list price $£ 8 / 17 / 6$, our price $84 / 96$, post and packing 10/-.

Three speed antomatic changer by a very famons manufacturet, current model, Will take 7 in ., 10 in . or 12 in , records mixed. Turnover erystal head.
A.C. Mains $200 / 250 . E / 19 / 6$. P. \& P. $4 / 6$.

CUB onc-sixth h.p. A.O. 220/230 v. by Brook Motors. Reversible for continuous running, £4/9/6 Post and pky. 7/6.
Radiogram Chassis, 5 valve A.C./D.C. 3 wave band superhet $195 / 255$ v. $19 \cdot 49,200-550$ and
 P.M., £8/17/6. P. \& P. 5/-.

## 

CONSTRUCTOR'S PARCEL, medium and long wave A.C. mains 230/250 2-valve plus mela rectitier, comprising chassis $101 \times 45 \times 11$ in., 2 wave band scale, tuning condenser, wavechange switeh, volume control, heater trans, metal rectiter, 2 valves and v/holders, emoothing and bias condensers, resiowrs and amall condensers, and med!
wound 22/6. P. \& P. $2 / 6$ extra. Circult and point-to-point, $1 / 3$.
CONSTRUCTOR's PARCEL, comprising chassis $12 \frac{1}{6} \times 24 \mathrm{in}$, cad. plated, 18 gauge, $7 / \mathrm{h}$. I.F. and trans. oit-outs, back-plate, 2 supporting brackets, 3 wave-hand scale, new warelength stationonames. size of scale 11 x 4 in., drive, sp.. drum, 2 pulleya, pointer, 2 bulb holders, 0
 $250-0-25060$
P. \& P. $3 / 6$.
CR100 Coil packs in frat-class condition less oscillator sectlon, complete with 4 -gang tuning condenser, 19/6. P. \& P. 3/6
CR100 485 KC. IF.E, types 3, 4 and 5 and F.B.O., new condition, $7 / 6$ each. 485 Kc . Xtal for CR100, $12 / 6$.
4-Laug tuning condenser for CR100, 9/6.
POLISHING ATTACHMENT for electric drills. Quarter inch spindle, chromium plated Sin. brush, 3 polishiag olothe and one aheepskin mop mounted on a 310 . rubber cup., Post and pkg. 1,6. 12/6. Spare sheepskin mops, $2 / 6$ each.
POTATO AND VEGETABLE PEELER. By Ismous manulacturer. To suit models A 200 and A700. Capacity 4tlbs, complete with water pump. All alominium construction, white stoveenamelled finimh. Originally intended for adap
verted for hand operation, 39/6. P. \& P. 3--

## USED A.O, MAINS 5 VALVE, 3 WAVE-BAND SUPERHET OHASSIS

 tuving condenser, majns transformer, volume control with switch, hone control, 3 waveband coll pack (this is a completely detachable soil pack on вeparake small chassis), varlous small
40-WATT FLUORESCENT KIT, A.O. mains 230/240. Oomprising choke, power factor condo.
20 watt A.O. D. D. $200 / 250$ v. FLUORESCENT KIT comprisalng terpugh in white stoved anamel
inntah, two tube holders, starter and holder and barretar. Poat and peokiog 1/6, 12/6.

Mains Droppers. 0.3 manss., 460 ohm tapped 280 and $410,1 / 8 ; 0.2 \mathrm{amp} .777$ ohms, tapper at 100 ahms, vitreoun, $1 / 6$ 0.3 amps. 050 ohms, kapped 700 and
$825,2 / 8 ; \quad 0.2$ amp., 1,000 ohms $\begin{array}{ll}825,2 / 8 ; & 0.2 \text { amp., } 1,000 \text { ohros, } \\ \text { vitreous, tapped } 2 / 6 ; \text { vitreous, } 0.3\end{array}$ Vitreous, tapped 2/6; Vitreous, 0.3 . 760 tapped $680,640,600,3 / 6$. P. \& ' $\mathbf{P}$. on each 3d
T.V. Width Controls, $3 / 6$

PERSONAL SHOPPERS ONLT. 9i Enlarger, 17/6; 12in. $27 / 6$
Germanium Crystal Diode, 1/6, post paid Used gin. Tube with ion burn, $17 / 6$, post paid
Line O.P. Translormer in aluminium can mounted in rubber. 18/6.
Speaker Matching Unit on alurninlum chassis, $3-15$ ohmo reversible, $12 / 6$. Line and E.L.T. Transformer, $14 \mathbf{K v}$. using forrucart core, complete with shields U37 rectiffer winding, $35 /$ Ling and E FT Transiormerg, 35/ using ferrocart core, complete with built-In line and wldth control. Mounted on small mil-chassis. Overall size $4!\times 1$ in. EV51 rec. Finding, $27 / 6$. Scan coils, low line low impedance rame, complele with lowe transormer to match above, 27/6. P. \& P. 2/Line and E.H.T. Tranaformer, 9 Kv. ferrocart core, EY51, heater winding. complete with mean oolls and frame output transformer, and line and width contm, $£ 2 / 6 /-$ P. \& P. 3/-
As above, but complete with line and 1 rame blockiug transtormers, 5 Henry 250 wlo 380 mA A.C. ripple $20119 / 6$ P. \& P. 3J-.

Vaive Holders, moulded octal Mazda and loctar, 7d. each. Paxolin, octal M7G. B8A and B9A, 7d. each. B7G moulded and B9A with screening can 1/6 each.
32 mifd 350 wkg
$16 \times 24,350 \mathrm{wkR}$.
4 mid., 200 wilg
$16 \times 8 \mathrm{mid}$., 500 wikg.
$16 \times 16$ mid., 500 wkg.
$32 \times 32 \mathrm{midd}$., 350 wkg..
${ }_{25}^{250} \mathrm{mfd} ., 25 \mathrm{wkg}$
250 midd., 12 w. wk.....
16 mid., 500 wkg., wise ende. 8 midd., 500 V. wkg., wire eads. 8 mfd ., 350 \%. wkg., tag ends.
$50 \mathrm{mid} ., 25$
\%. wkg., wire ende. 100 mid., 350 wkg., wire ende. 100 mid., 450 w. wkg., 280 mA . A.C. 150 mipple . 350 . wkg ., 280 mA . 200 med., 275
$200 \mathrm{mkd} ., 2754 \mathrm{~kg}$. . $16+16$ mfd., 350 wkg
$16+16 \mathrm{mfd},{ }^{350} \mathrm{wkg}$
50
$\mathrm{mfd} .$,
180
wkg
65 mid., 2200 wkg.
8 mid., 150 wkg
$60+100$ midd. 280 wkg
30 mfd., 12 wkg.
Mininture wire ends moulded,
500 pl., and . 001 , each, 7 d.
Combined 12 in mank and escutcheon, in lightly tinted Perspex. New sspect edged in brown. Fits on front of $17 / 6$.
Frame Osoillator Blocking Trans., 4/6 CHOKES: 2.20 Hen. 150 mA ., $15 / \%$ P. \& P. $3 /-$
6 Hen., 275 mAA .15 L,
100 Hea. $40 \mathrm{~mA}, 15 /=$ P. \& P. $3 /-$ 2 henry $150 \mathrm{~mA}, 3 / 6: 250 \mathrm{~mA} .10$ heary 10/6; 5 henry $250 \mathrm{mA}$.60 ohms.

Wide Angle P.M. Foons Units. Vernier adj. state tube, 15/-
Energised Focus Coil, fow resistance mounting bracket, $17 / 6$.
Ion Traps for Mallard or English Electric tubes, 5/-, post paid.
Standard 465 Kc . iron-cored I.F., $4 \times 1+\times 1$ inin., per pr, \%/8. Wearite $3 \frac{1}{2} \times 1 \times 1$ ilin., per pr.s $9 / 6$.
Lron-Cored 465 Kc. Whintle Filter, 2/6 485 KC . MIDGET I.F.s, Q. 120 bize $1 \frac{1}{2} \mathrm{in}$. long, 1 ln . wide, tin. deep by very famous manutacturer. Pre-aligned adjustable iron-dust cores, per pair,
12/6. $12 / 6$.

SIGNAL GENERATOR
Coverage $120 \mathrm{Kc} / 5-84 \mathrm{Mc} / \mathrm{s}$, £4-19-6
P. \& P. 4/-.

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dors from reeeritt of order. Where post and pocking charge is not enquiries S.A.E., lists Sd. each.

PATTERN GENERATOR
Coverage $40 \mathrm{Mc} / \mathrm{s}-70 \mathrm{Mc} / \mathrm{s}$, £3-19-6
P. \& P. 4/.

VALVES-SPECIAL BARGAINS IN GOVT. SURPLUS
 VALVE ORDERS OVER €S, less $5 \%$ discount, post free.
R.C.A. SPEECH AMPLIFIER MI-1I220A. 200250 v. A.C. i Containing 2-6L6 valves in push-pull ouspur, 7 valves in all. Gram and mikeinputs fitted. This model is undoubtedly one of the finest medium powered amplifiers ever produced. Complece with all valves, $\mathbf{1 8} 18 / 10 /-$ Carr. 10/-.
50 WATT AMPLIFIER, EX-GOVT. With 4-KT66's in paralleled push-pull, Standard 200-250 v. mains inpu\&, A.C. Output impedance 600 ohms line. For high imp. gram. and mike input. Bass boosc control fiteed. This excellent quallty amplifier is housed in a strone metal case and is ready for use. Terrific performance. Bargain value $\mathbf{E 2 5}$, carr. Daid HALLICRAFTER SPEECH AMPLIFIER BC.614A. Part of BC6IO equipment. As new, $\mathbf{\infty 2 5}$, carriage paid.
VITAVOX PRESSURE UNITS COMPLETE WITH 42in. EX PONENTIAL HORN, $\boldsymbol{\epsilon}^{7} / 12 / 6$, carriage $10 /$. Pressure Unit can be supplied separately at $\varepsilon 4 / 9 / 6$, carriage $5 /$ -
ROTARY CONVERTERS. 24 v . D.C. input, 230 V . A.C. output at 50 cycles, 100 watt. Brand new, $£ 4 / 17 / 6$, carriage $7 / 6$.
VARIABLE VOLTAGE REGULATOR TRANSFORMERS. Input 230 v. A.C. at 21 amps. Output 57.5 in 16 equal steps 10230 v at 21 amps. Ex-Govt. In perfect condition. 112 , carriage 101 -
LATEST MODEL MONARCH B.S.R. 3-SPEED CHANGER, 69/7/6, carriage paid.
RA-88 RECTIFIER UNIT (PART OF SCR. 720 EQUIPMENT). Containing the following valve line-up: $3-6 \mathrm{~L} 6 \mathrm{~m}, 3-5 \mathrm{~T} 4 \mathrm{~m}, 2-6 \mathrm{SL} 7 \mathrm{gt}$ 's 2-VR. $150 / 30$ 's, etc., etc. Bargain value at $64 / 19 / 6$, carriage 5
CRYSTAL CALIBRATOR BY MARCONI INSTRUMENTS.
Brand new and unused, complete with spare set of 5 valves and operating manual. Freq, range $170-240 \mathrm{Mc} / \mathrm{s}$. Accuracy: I part in 10,000 . ing manual. rreq. range $10 / 0$
AMERICAN H.F. SIGNAL GENERATORS. Type 122A. Input 110 v. A.C. Freq. range: $8-150 \mathrm{Mc} / \mathrm{s}$ and $50-230 \mathrm{Mc} / \mathrm{s}$. New and unused $\$ 10$ v. A.C. Freq.
E25, carriage $10 /-\dot{C L}$ SLOT METERS. $200-250$ v. A.C. at S - 10 amps. I/-inslotand 6d. or 7d. perunit. By Measurement Ltd. All bakelite case In very good condition, $50 /$., p. p. $2 / 6$.
R. 1155 COMMUNICATION RECEIVERS, Individually tested and despatched in good working order. Cases slightly soiled, $£ 8 / 19 / 6$. Brand new, $£ 10 / 19 / 6$, carrlage $7 / 6$.
RECEIVER TYPE 109. In good condition. Freq, range $1.8-3.9 \mathrm{Mc} / \mathrm{s}$ and 3.9-8.5 Mc/s continuous, Designed to operate on 6 v . battery. Limited quantity only, $£ 4 / 7 / 6$, carriage $10 /-$
PHOTO-ELECTRIC MULTIPLIER CELLS. Type 931 . $22 / 10 /-$ p.p. 1/-. Also 931A complete on chassis with mulciplier network and 2-832 valveholders, etc., $£ 3 / 10 /-$, p.p. $2 /$.
TIME SWITCHES (NEWBRIDGE). 250 v. A.C. Synchronous 5 amp. Used but in good working order, 39/6, p.p. 2/6.
CONDENSERS, 2 mfd .7 .5 kv , working at is kv. Test (Dubilier) Brand new, 62/10/- each, carr, 5:
Brand new, ©2/0/-each, carr, Si-̇ (EX-G,P.O.) HIGH RANGE CONSTANT PRESSURE MEGGERS. 5 meg. 1,000 meg. infinity. in good working condicion. Housed in wopden case, $\mathbf{2 7 / 1 0 / - , \mathrm { Garr } , 1 0 / \text { - }}$ In good working condicion. Housed in wopden case, $27 / 10 /-$, Garr. $10 /-$
Ditco, 250 v . Low range bridge megger. 5000 ohms- 20 megs. infinity Ditto, 250 v . Low range bridge megge
Finish as above, same price and carr,
813 CERAMIC' VALVE HOLDERS, $9 / 6$ each, p.p. 6d, Also 4 pin large Jumbo ceramic valve holders, $6 / 8$ each, p.p. 6d.
TRANSMITTER-RECEIVER T.R. II43A or 1430. Either type uvailable. Complete Unit (less crystals and power supply). In very good condition. This irans-receiver, which until recently was extensively used by the R.A.F., has a very wide freq, range and is offered at the ridiculously low price of $49 / 19 / 6$, carr. IO/-.
CATHODE RAY TUBES. Types 3BPI, 3 in, , new and unused with base and screen, 42/6, p.p. 2/-. Type VCR138 (ECR 35), $3 \frac{1}{2}$ in, with screen and base in new and unused condition, 42/6, p.p. 2/-. Type VCR97, 6in., ex-equip., in good order, 20/-, p.p. 3/6. VCR131, lin. C.R.T., new and in perfect condition. Miniscope replacement tube, etc., 35\% p.p. 1/-. Trpe CVI526, 27 in., 4 v . filament, $3,000 \mathrm{v}$. anode, complete with base and Mu metal screen, $20 /$. p.p. $2 / 6$.
MASTER CONTACTOR MK, II. REF, IOA!10994. 24 v . new and
unused, $10 /$-. p.p. 2/NDICATOR UNIT (EX-U.S.A.). Containing I.3BPI 3 in. C.R.T., $2-6 S N 7 s, 2-6 \mathrm{H} 6 \mathrm{~s}$, l-6G6, $1-6 \times 5,1-2 \times 2,7$ valves in all. Ideal for 'Scope conversion. New in original sealed cartons. Bar-5/- carr. 5/
DRY BATTERIES (EX-GOVT.), 72 y. H.T. I. 5 v. L.T. Ideal for battery portables, size: $6 \frac{1}{2} \mathrm{in}$. $\times 5 \mathrm{in}, \times 2 \mathrm{in}$. In very good condision cested before despatch. Price 5/-carr. paic.
TELESCOPIC DINGHYAERIALS. Min. Iengih I4in., max. length $7 \mathrm{ft} .4 / 6$ each, p.p. 9 d .
SELENIUM METAL RECTIFIERS BUILT TO SPECIFICATION. ALL TYPES AVAILABLE FULLY GUARANTEED, VERY GOOD DELIVERY. COMPETITIVE PRICES. MAY WE QUOTE YOU? K3/ AND K8 SERIES STILL AVAILABLE FROM STOCK.
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Horn loaded....an-1,000 c/s cut off Equivalent moving-coll $-\frac{3.00}{5 \times 10^{5}} \frac{\mathrm{~ms}}{}$ units $\int \times 10^{5}$ dynes $/ \mathrm{gm}$ Dimenslons …....... $8 \frac{1}{2}^{\frac{1}{2}} \times 5 \frac{1^{\circ}}{} \times 4 \frac{1^{\frac{1}{2}}}{}$ Weight
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Wire Wound Controly R.C.A. Colver 0 ohms. Colvern CLP1232/268 500 ohms. Colvert 1,000 ohma, ppindle 1 In .

HEATER TRANSFORMERS 230 v . Imput '2 volt . 5 amp amp $4 / 6$ 230 v . Input 2 volt 3.0 amp .

 230 v . Imput 6.3 volt 1.6 amp . $6 / \%$ 230 v. Input 6.3 volt 3.0 amp. $9 /-$
230 v. Input 12 volt .75 amg.

SENTERCEL RECTIFIERS RM1, 3/9 еа; RM9, $4 / 2$ ез.; RM3 METAL RECTIFIERS
 $4 / 6$ еа.: 2 v. 1 sпир. $3 /-$ еа.; 250 v


3 Valve ( $6 \mathrm{~K} 7,6 \mathrm{~J} 7,6 \mathrm{~V} 6 \mathrm{GT}$ ) Plus Metal Rectifler, $2_{\text {2 }}$ Wave Bhad Midget radio AC Malns, complete
in every detail, full instructions, circuit diagram and shopping list 1/6. Complete kit down to nuts. bolts and solder f5/10/-. Post 2/6

POCEET TEST METER Ex-Covt. volt meter two ranges

## IRON ELEMENTS

 Standard adaptable type, 230 v. v . replacement type, $3 / 9$ ea H.M.V. replacement type $31-$ ea.PLIERS, with side cutters, $4 / 3$ pair.
PUBLICATIONS Each
Midget
Manus 1 Radio Construction
 Rape recorder Oncillowcope
Radlo Calculations Manual Practical Circulta Manual Radlo and Television Laborators Manual
L.E. Mb Silver Mlea, $1,000 \mathrm{pF}$
L. E.N. Silver Mica 100 pF . $3 \%$. 3 .d
T.C.C. Nllver Mics 50 pF . $10 \%$. 3 id.
L.E.M. Sliver Mica $300 \mathrm{pF}, 10 \%$ 3td.
L. E. M. Sllver Mica $350 \mathrm{pF} .10 \%$. 3 3d.
L.E.S. Silver Mlea $25 \mathrm{pF} .20 \%$
T.C.C. Silver Mica $1,000 \mathrm{pF} .10 \%$
3 3d

Tubular Condenser .s mfd. 500 v. 6 m Yruley Ewitch 1 pole 8 way 1 tim spindle
Arnow Toxgle slotted Dolly s.P. 1/9
Single screemed Cable ...... yard 6 d .
Eric Malna Dropper 725 ohms $71 \%$ tapped
Dubiller Moulded Mica A P9 . .005 $\frac{1 / 6}{4 / d .}$ Gibrater Wound Choke
Dubller Nitrogol 84512 mid 50 ${ }_{\text {3nternational Octai valve Holders }}$ 5/6 Paxolla …................... 41 d . Vitreous Enamelled Resistors
B.8.R. MONARCH AUTOMATIC These units will autochange Cn : They play MIXED 7 in ., 10 in . an 12ia. records.
They have separtive sapphire for L.P. and 70 r.p.m. which are moved into position by a simple switch.
Minimum baseboard size required
$14 \mathrm{in} . \times 12$ in., with height ubowe $14 \ln . \times 12$ in., with herght ubove 2jin. A butk purchase enables us to otter these BRAND NEW UNITS at
this exceptional price. These units this exceptional price. These units
are beantifully Anished in cream ename! with cream bakelite arm.
COMPLPTE WITE FULL. IN. STRUCTIONS. E9/19/6.

AMERICAN INDICATOR UNIT Brand new Incorporating 3in. tube B8P7 ${ }^{3}$ Wit 2 ghact BG6G, 9 patontiometers, 24 \%. nerial switch motor, tranaformer, and a hoet of small compouents. The whole unit which measures only 8 inn. $x$ $83 \mathrm{in} . \times 13 \mathrm{in}$. is brand new, enclosed in hlack crackle box, and can be
supplied at $65 /-$ plus $\$ /=\mathbf{P}$. \& $\mathbf{P}$.

Standard 1!ia. Brown Knobs, per
Zenith Dropper 910 ohms. each
Bakelite case. Double coll Buz-
zera, each.
Erie Dropper, $1, \$ 40$ ohms, 150 ohms, each Bon. 4 BA Nuts and Bolta, each $\ldots . . . .$.
Intervalve Transformers Ex. Equip, each Hand Microphone Bakelfte awitch in handle, each High to low Impedance headphone units (Insert in lead) each
Rubber grommets, assorted, per dozen Bakelite Needle Cups, each
leeving, various colours, 1 MM , lenath
Yains 8witch, 2 hole fixing, S.P., each
Yaxley Switch, 1 Poie, 9 Way, ench
Wased Carton, 8 MFD., 430 v. , each
Westectors, WX12, W4, WX6, each
Collaro Hl Fl Pickup, each
TERMS: Cash with order or C.O.D. Post ${ }^{-}$ age and Packing charges extra, as follows: Orders value $10 /$ add 9 d .; $20 /$ - add $1 / \%$ 40/- add $1 / 6$; 65 add $2 /$ unless otherwise stated. Minimum C.O.D. fee and post age $2 / 3$.
MAIL ORDER ONLY

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3-WAY MOUNTING TYPE
PTimary: 200-220-240
$0-5$ v. 2 amp. Both tapped at 4 . $0-6.3$ v. 4 amp.
MT2
Primary: 200.220-240 v
$0-5$ v.
PENCL RECTIFIERS
К3;25 5/8: K3/40, 7/6: К3/45, 8/2; K3/50. 8/8; К3/60,.8/8.

| CONDENSER3 |  |  |
| :---: | :---: | :---: |
|  | 'Each | Each |
| $8 \times 8 \mathrm{mld} .450$ v. | 4/- | $32 \times 32 \mathrm{mfd} .450$ - . . . 6/11 |
| $8 \times 16 \mathrm{mfd} .450 \mathrm{v}$. | 4/0 | $32 \times 32 \times 8 \mathrm{mld} .350$ 7. 5/6 |
| $8 \times 24 \mathrm{mfd} .350 \mathrm{v}$. | 3/- | $92 \times 32 \mathrm{mfd}$. 350 v . |
| $8 \times 32 \mathrm{mid} .475 \mathrm{\nabla}$ | 3/8 | 45 mid .25 v. ...... $5 /$ |
| $12 \times 4 \mathrm{mfd} .450 \mathrm{v}$. | 2/* | 60 mid. 430 v. . . . . . . . $2 / 9$ |
| $15 \mathrm{mid}$.450 จ. | 3/- | 64 mid. 350 v. . . . . . . . . $2 /$. |
| $16 \times 8 \mathrm{mfd} .350$ จ. | 4/= | Dubilier (B.R. Range): |
| $16 \times 16 \mathrm{~m} / \mathrm{d} .350$ จ | $3 / 3$ | BR 850. 8 mid .500 7. 2/9 |
| $16 \times 16 \times 8$ mfd. 350 | $3 / 6$ | BR. 1650.16 mfd .500 v. $3 / 3$ |
| $20 \times 20 \mathrm{mld} .500 \mathrm{v}$. | 4/8 | BR.2050. 20 mfd. 500 จ. $3 / 6$ |
| 24 mid .450 V. ${ }^{2}$ | 219 | $8 \times 8 \mathrm{mfd} .300 \mathrm{v} . . . . . .4 / \%$ |
| $24 \times 16 \mathrm{mid} .350 \nabla .$ <br> 32 mild .450 v. |  | BE.501. $50 \mathrm{mfd} .12 \mathrm{v} . . .1 / 9$ |
| $32 \times 8 \mathrm{mfd} 350 \mathrm{v}$. | $3 / 6$ | $16 \times 16 \mathrm{~mm}$ (d. 500 v. .... 5/- |
| $32 \times 16 \mathrm{mfd} .350 \mathrm{v}$ | $4 / 6$ | $16 \times 8$ mid. 500 v. ... 4/9 |

OSMOR COIL PACKS
тхре H.0., 48/- each. Type L.M., 40/= each. Typa T.B., $50 /-$
SPRAGUE CONDENSERS
$.05 \mathrm{mfl} ., 500$ v.; $0.1 \mathrm{mid} ., 1,000$ v.; $1 \mathrm{mfd} ., 350$ v.; . 02 mfd . 750 v. All $9 /$-doz.


- This attractive walaut finished cabinet is avallable t for 6/1n. or 8in. speaker units. Metal speaker fret.
6!in. type: Meanure 8tin. $\times 8 \$ 1 \mathrm{n} . \times 4 \mathrm{fin}$. at base t

$\star \star \star \star \star \star \star \star \star$

Rola Sin. Bpeaker
Hoodmans 5In, unit
LOUDSPEAKER UNITS

Plesser 6|la lightwetght uni
Rois 61 ln . standard type
bectrona 6 fin. With transtormer
Truyox 6 in. wafer type
Truvox 6 fin . Wafer type
leasey $81 \mathrm{~m}_{\text {. }}$ lightwelght unit
Elliptical 4ia. $\times 7 \mathrm{in}$. unit
Malna energised 81 n unit, $1,000 \mathrm{a}$
mains euergised bin. unlt, 600 D
$2 / 6$

$2 / 6$ Cormert. Available tin the following 6in $\times 4 \operatorname{in} \times 2 \operatorname{lin}^{\text {sin. }} \times 6 . . . . . .$. $81 \mathrm{n}, \times 6 \mathrm{in}, \times 24 \mathrm{n}$. | $101 \mathrm{n} . \times 7 \mathrm{hn} . \times 24 \mathrm{in}$ |
| :--- |
| $121 \mathrm{n} . \times 81 \mathrm{n} . \times 24 \mathrm{n}$ | $14 \mathrm{in} . \times 8 \mathrm{in} . \times 2$ in.



613 ea
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12in. type, $11 / 6$ ea: $14 / 6$ ea.

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Thi popular range is suitable
This popular range to suitable for an Televielon constructors, etc. Keep your costs down when bulling the Argua' or "Simplex recelvens, typer $2 \mathrm{KO}, 5 \mathrm{KO}, 10 \mathrm{KO}, 20 \mathrm{KO}$ $25 \mathrm{KO}, 50 \mathrm{KQ}, 200 \mathrm{KO}, 100 \mathrm{KO}$ :


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$4 \frac{1}{2} v$ Heavy Duty Bell Battery. Size $6 \frac{1}{} \times 4 \frac{1}{2} \times 2$ ininies
$4 \frac{1}{\frac{1}{v} \text { v. Heavy Duty Bell Battery. Size } 6 \frac{1}{2} \times 4 \frac{1}{2} \times 2 \frac{1}{2} \text { in. }}$
72 v . H.T. 1.5 v . L.T. Size $6 \times 5 \times 1 \frac{3}{\text { in in. }}$
150 y . H.T. Size $2 \frac{2}{2} \times 5 \frac{1}{2} \times 1 \mathrm{I}_{\mathrm{i}}^{2} \mathrm{n}$.
$4 / 6$
$5 / 6$

All batteries sealed and unused. All plus $1 / 6$ post and pkg. Special
reduction for quantitles.
ELECTROLYTIC CONDENSERS Per doz.
16 mid. $375 \quad \mathrm{r}, 2 /=$ each
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$20 \times 20 \mathrm{mfd} .275 \mathrm{v} .2 / 3$ each
24 rafd. $350 \mathrm{v} .1 / 6 \mathrm{each}$
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BIAS CONDENSERS
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4 mfd. 400 v. D.C. $3 / 6$ each. Many other types in stock. Your enquiries invited.
LARGE ASSORTMENT OF TUBULAR CONDENSERS
MIDGET MICA CONDENSERS. .0001,.0002,.0003,.0004, .0005 5/-
200 Assorted Moulded Micas. Popular Values ....................... $<2$ 10. 0 200 Assorted Silver Micas. Popular Values
200 Assorted Carbon Resistors: t t and I watt. Goo...................... $\in 2$ selection 100 200 Assorted Carbon Resistors: $\ddagger . \frac{1}{2}$ and I watt. Good selection \&1 10
0.1 mfd. 12,000 volts test Mansbridge Condensers. Height $6 \frac{1}{3} \mathrm{in}$. Width $3 \frac{1}{2} i n$. Depth $2 \frac{1}{2} i n$. Fixing Centres 4in. Plus $/ /$-post ... $5 / 6$
PAXOLIN SHEET
$18 \times 4 \frac{1}{2} \times 1 / 16 \mathrm{in} . \mathrm{I} / \mathrm{-}$ each; $10 \times 10 \times 1 / 16 \mathrm{in}$. . $1 / 6$ each; $20 \times 10 \times 1 / 32 \mathrm{in}$. . $1 / 6$ each; $20 \times 10 \times 1 / 16 i n_{n}, 3 /-$ each.
RESISTORS
Carbon $\frac{1}{2}$ watt $2 / 6$; $\frac{1}{2}$ watt $3 /=$; I watt $4 /=; 2$ watt $6 /-$ per doz.
WIRE WOUND AND VITREOUS. 5 watt $1 / 6$; 10 watt $2 / 6$; 15 watt 3/-: 20 watt $3 / 6$ each.
HIGH STABILITY. $\frac{1}{2}$ watt $5 \% 6 \mathrm{~d} . i \frac{1}{2}$ watt $5 \% 9 \mathrm{~d}$. i I wate $5 \% 1 / 3$ each. HIGH STABILITY. $\frac{1}{2}$ watt $5 \%$ 6d. i $\frac{1}{2}$ w
A few values in $1 \%$ and $2 \%$ still available.
A few values in $1 \%$ and $2 \%$ still available. PLEASE AS WE CANNOT ALL ORDERS FOR RESISTORS C.O.D.
GUARANTEE TO STOCK ALL VALUES.
GUARANTEE TO STOCK ALL VALUES.
W.W. V/CONTROLS. ALL WELL-KNOWN MAKE5. Pre-set $2 / 6$ each W.W. V/CONTROLS. ALL WELL-KNOWN M
Spindle types $3 /$ each. Values from 5 ohms to 50 k .

V/CONTROLS WITH SWITCH 5k, 50k, $\frac{1}{2}$ meg., 1 meg. ... $3 / 6$ each V/Controls Less Switch. Most values spindle and preset ....... 2/- each
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2T/-doz.
Push Button Knobs
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TAG STRIPS. 3-way $2 /-$; 4-way $2 / 6 ; 5$-way $3 /=; 7$-way $4 /-; 28$-way $12 /$ - doz. SLEEVING. $2 \mathrm{~mm} .2 / 6 ; 3 \mathrm{~mm}, 3 / 6 ; 4 \mathrm{~mm} .4 / 6 ; 5 \mathrm{~mm} . . . . .5 / 6$ per doz. yd.
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VALVE HOLDER FITTED WITH LOWWR CAN I/6 per doz. extra
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Paxolin $V / H$ Ine Oct. B9A B7G
$4 / 6 \mathrm{doz}$.
6/- doz.
7-pin..
STANDARD 5CREENING CANS 3-piece I/-each; Spring Loaded
BELLING LEE PLUGS AND SOCKETS, 5 -pin $1 / 9 ; 7$-pin $2 /=$;
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IR SPACED TRIMMERS 5, 10, 15, 20, 25, 50 and 75 of pre-
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GROMMETS I grs. assorted grommets $\frac{1}{\text { tin. to lin. }}$
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18/- doz.
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TRANSMITTER UNIT Ex-TR1143A.-Suitable for conversion to 2 metres. Circuit diagram and coil conversion details supplied free. Price, less valves, $5 /$ - post paid.
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4 henries 40 mA (Potted), $5 / 6$ post paid. 8 henries 100 mA (Potted), $8 / 6$ post paid.


## Miniature POCKET RADIO

Incorporating high "Q" technique using the New Ferrite rod. Made possible by simple conversion of an ex-Govt. Hearing Aid.
Technical Details. A Germanium Diode Detector circuit followed by the existing 3 valve Amplifier, giving adequate amplification throughout the medium wave band.
This conversion can be carried out in approximately 30 minutes.

THE COMPLETE KIT OF PARTS includes a Type OL10 Hearing Aid (with Crystal microphone) in perfect working order with miniature ear phone and moulded ear insert attached: ferrite rod, germanium diode, components, circuit diagram and full instructions. Price £2 6s.0d. post paid (or without crystal microphone $£ 22 \mathrm{~s}$.). Batteries extra: 1.5v. L.T. (Type D18), 8d.; 30v. H.T. (Type B119), 4/3.
NOTE: As the crystal microphone is not used in the Pocket Radio, it can, if desired, be used as a general microphone and it does not require a matching transformer.


SEE and HEAR this Miniature POCKET RADIO demonstrated.


#### Abstract

DROPPERS. Chassis type heat dissipating droppers for midget receivers. $5000 \Omega 10 \mathrm{~W} .3 / 6$ post paid. MAIN DROPPERS. $7500 \Omega 50 \mathrm{~W} .2 / 6$ post paid. NEON TUBES. Miniature bayonet type. 80 V striking ( 1 megohm in series for mains). $1 / 6$ post paid. ABSORPTION WAVE METER. Easily converted to 2 metres or 70 cm . In Copper-plated metal case 3 lin . $\times 4 \mathrm{lin}$. $\times 5 \frac{\mathrm{in}}{}$. with dial calibrated $0-100$ and 80 V Neon Tube. Coverage approx. $190-210 \mathrm{Mc} / \mathrm{s}$. New. $6 / 6$ each post paid. 25 WAY 2 'BANK SWITCH. Contained in metal case $9 \mathrm{in} . \times 6 \mathrm{in} . \times$ 2 lin. with two 12 way terminal blocks. Useful for multi-test meters, model control, etc. Type 5D/460, $7 / 6 \mathrm{pqp}$.


## BENDIX COMMUNICATIONS RECEIVER TYPE RA-10DB

A superb 8-valve 4 band receiver covering $150-400 \mathrm{kc} / \mathrm{s}, 400-1100$ $\mathrm{ke} / \mathrm{s} ., 2-5 \mathrm{Mc} / \mathrm{s}$, and $5-10 \mathrm{Mc} / \mathrm{s}$. Valve line up $6 \mathrm{SK} 7 \mathrm{R} / \mathrm{F}$, $6 \mathrm{~K} 8 \mathrm{~F} / \mathrm{C}$, Two 6SK7 IF Amplifiers, 6R7 Second Det. AVC and AF Amplifier, $6 \mathrm{C} 5 \mathrm{BFO}, 6 \mathrm{~K} 6 \mathrm{OP}, 6 \mathrm{H} 6 \mathrm{Sig}$. limiter diode.
SENSITIVITY. 4 mV . signal, gives 50 mW . audio at 4-1 signal/noise ratio. SELECTIVITY. At $5 \mathrm{Mc} / \mathrm{s}$, is $39 \mathrm{kc} / \mathrm{s}$. at 1,000 times down. At $150 \mathrm{ke} / \mathrm{s}$, bandwidth is only $22 \mathrm{kc} / \mathrm{s}$. IMAGE RATIO. Minimum image ratio is better than $1500-1$ on $5-10 \mathrm{Mc} / \mathrm{s}$ Band and $10,000-1$ on $150-400 \mathrm{kc} / \mathrm{s}$.

As a BOAT, TRUCK, CARAVAN or CAR RECEIVER it is UNEQUALLED in value; converted to a.c. operation for fixed station, it equals receivers selling for over five times the price we ask.

PRICE
£5. 10.0
Pkg. \& Carr. 10)-.

## MULTIRANGE A.C./D.C. TESTMETER

## of well known American manufacture

This testmeter has a basic movement of 400 microamps and is calibrated for use on the following ranges:-
A.C. and D.C. Volts 0 to $5,000 \mathrm{~V}$ in 6 switched ranges.
D.C. Current ranges $0-1 \mathrm{~mA}, 1-10 \mathrm{~mA}, 10-100 \mathrm{~mA}$ and $100 \mathrm{~mA}-1 \mathrm{~A}$. Use as an OHMETER (Resistance Measurements). $1 \Omega$ to 1 Megohm. Decibels from -10 db to +15 db . For line load impedances from 5 to $1,000 \Omega$ (directly calibrated for $500 \Omega$ line).
This instrument is contained in a well finished polished wood case with leather carrying handle. Leads and test probes are housed in the case which measures $6 \frac{1}{2}$ in. $x$ $6 \frac{1}{2}$ in. $\times 4 \frac{1}{2}$ in.
All meters fully tested before despatch. Supplied complete with moulded test probes, full operating instructions and circuit diagram. PRICE
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Packing and Postage 3/.

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Telephone: LANgham 0141
3 WAVEBANDS: - L.W. $800 \mathrm{~m}-2000 \mathrm{~m}$, M.W. $200 \mathrm{~m} \cdot 550 \mathrm{~m}$, S.W. $16 \mathrm{~m}-50 \mathrm{~m}$
Chassis size $13 \frac{1}{2} \mathrm{in}$. $\times 54 \mathrm{in}$. $\times 2 \not 2 \mathrm{in}$. Attractive Glass Disl $10 \mathrm{in}, \times 41 \mathrm{in}$. edge Lit by
2 pilot lamps, Horizontal or Vertieal Station Names and 4 control knohs, walnut or
ivors to choice. ${ }^{4}$ position W/C switeh, L.M.S. and Gram. P.U. sockets. Miodern
circuitry, sll coils ajjustable dust cored and only quality components used throurhout
Delaged A.V.C. and nea. feed-bnek. A.C. mains $200 / 250$ F. Doublo wound transf
solates chassis irom mains. Aligned and eslibrated ready tor use.
BRAND NEW \& GUARANTEED 59.15 .0 Carr. and ins. $4 / 6$.
3 -ohm speakers suitable for this chassis available $8^{\prime \prime} 17,610^{\prime \prime} 25^{\prime \prime}=$
This chassis is a genuine bargain and delivery is reasonably good.

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> Latest P.S.R. Model. 3 speed Autochanger Mizer Unit. Farnous Magidisc 7in, 10 in , and 12in. recod seleetor. Modern cream strling. Dual Xtal cartridge stylus for high Adelity reproduetion. Ag used by leading Radiogram Manufacturers. Complete with fall inntractions snd template. OUR BARGAIN PRICE $9 \frac{1}{2}$ gres. post free.

## JASON F.M. TUNER UNIT $87-105 \mathrm{mc} / \mathrm{s}$

Kit of parts to build this modern and higbly suceessfal unit complete with drilled chassis and J.B. dial, wound coils and gereening eans, 4 B. V.A. miniature valves, and brated mo/s., edge lit by 2 pllot lamps, $12 / 6$ exira. Powbr Pack components kit ineluding double wound Mains Transiormer eq ejpextra Tented and approved by "Radio Constructor," etc. Flastrated handbook with full details $2 / \mathrm{F}$, poitt tree.

ELECTROLYTICS Leading Makes New Stack


16
80 cable ${ }^{\circ}$ CO-AXIAL SPECIAL. Semi-air spaced polythene, 90, 12 yds. $8 / 9$.
COAX PLÚAS
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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Lieutenant | ... | ... |  | . | 23 | £686 | ¢823 |
| Lieutenant | ... |  | ... | ... | 25 | ¢723 | 8860 |
| Captain | ... | ... | ... | ... | 27 | 6866 | ¢1,005 |
| Major | ... | ... | ... |  | 34 | [1,193 | ¢1,330 |
| Lt.-Colonel | ... | ... | ... |  | 42 | ¢1,512 | [1,640 |
| Colonel | ... |  | ... | .. | 45 | ¢1,795 | [1,922 |
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[4969
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[0262
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sionable and offer considerable scope for sionable and offer considerable scope for

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Vacancies exist in new, well equipped and expanding laboratories of the Chemical
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3. SENIOR ENGINEER-for work on general circuit development, with sound fundamental knowledge of electronics and the ability to apply it.
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August. 1955 . August. 1955.
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[^0]:    October: Show Review. An analysis of design trends in television and sound broadcast receivers.

[^1]:    * Langford-Smith, F., "Radiotron Designer's Handbook", 3rd ed. 1940

[^2]:    * Mu!lard Valve Measurement and Application Labcratory

[^3]:    *See Wireless World, April, 1951, for principles and description of early type.

[^4]:    * "The Human Use of Human Beinas," 'evised edition 1954 Eyre and Spottiswoode.

[^5]:    † "Electronic Positioning," Wireless World, January, 1955.

[^6]:    * Acoustical Manufacturing Co. Ltd.

[^7]:    * See for example "Colour Television on 405 Lines" Wireless World. June, 1954

[^8]:    * "Colour Television on 405 Lines," Wireless World, June, 1954.

[^9]:    * "Electronic Voltage Stabilisers," by Hunt and Hickman, Rev. Sci. Inst. Jan. 1939, p. 6.
    †" A Low-noise Amplifier," by Wallman, Mannee and Godsden Proc. I.R.E., June 1948, p. 700.
    $\ddagger$ " Cascode Audio Amplifier has Low Noise Level," by R. L. Price, Electronics, March 1954, p. 156.

[^10]:    * Noise voltages, like a.c. voltages of unequal frequency, have to be added together in this way; see" Total Power" in the March 1952 issue.

[^11]:    *" A Cascode Amplifier Degenerative Stabilizer," Electronic Engineering, April 1955, p. 174.

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