# Wireless Worlal 

MARCH 1954 • TWO SHILLING S

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VALVES AND GATHODE RAY TUBES
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## Wirceless TJorld

RADIO, TELEVISION<br>AND ELECTRONICS

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## 15. VALVES FOR V.H.F. TELEVISION RECEPTION

The advent of television transmissions in the frequency range of 174 to $216 \mathrm{Mc} / \mathrm{s}$ (Band III) will mean that receivers must be capable of receiving signals radiated at these frequencies in addition to those already transmitted in the range 41 to $68 \mathrm{Mc} / \mathrm{s}$ (Band I).
Reception of signals at the higher frequencies involves changes in the input and frequency changer stages of present-day superheterodyne receiver designs. To meet these requirements the Mullard "World Series" of television valves has been augmented by the introduction of two new types:-

PCC84-Double triode for use as a cascode amplifier in V.H.F. input stages.
PCF80-Triode pentode for use in frequency-changer stages.
As with most of the other "World Series "valves, these new types use the B9A (noval) miniature nine-pin base, have heaters suitable for inclusion in a 300 mA series chain, and can operate with a receiver h.t. supply of 180 volts.

PCC84. The double triode is designed for use as a cascode amplifier, i.e. one triode section operates as a neutralised grounded-cathode amplifier and the other section as a grounded-grid amplifier. The two sections are then connected in series across the h.t. supply, the anode of the first section having a direct connection to the cathode of the second section. The cascode arrangement results in a low noise level for the input stage being achieved in the first section, combined with high gain resulting from use of the grounded-grid connection in the second section.

PCF80. The triode pentode, having separate electrode systems for triode and pentode sections, is used in the frequency changer stage. The triode section, connected as a local oscillator, is in a Colpitts circuit. The oscillator voltage and the input signal from the PCC84 are mixed either indactively or capacitively on the control grid of tne pentode section. This section functions as the mixer stage, its performance depending considerably upon the choice of circuit components to give optimum values of conversion conductance and input damping.

The accompanying functional diagram illustrates the basic mode of operation of these valves when used in the " front-end " circuit of a television receiver for use in Bands I and III.


#  

INTEREST in proposals for the setting up of a v.h.f. broadcasting system may be stimulated by the announcement, made in Parliament on February 18th, that the Postmaster-General has accepted the recommendation of the Television Advisory Committee that frequency modulation should be used. Acceptance of this recommendation was accompanied by a statement that B.B.C. proposals for making a start on an initial group of stations are now under discussion with the Corporation and "further developments will be considered in due course."

This somewhat lukewarm acceptance of the T.A.C.'s Report* is perhaps on a par with the slightly unenthusiastic nature of the Report itself, which has provoked surprisingly little comment in radio circles since it was published on January 18th. The Report starts off on a rather doleful note by saying that the introduction of v.h.f. broadcasting is an "unwelcome complication." In addition to the question of modulation already mentioned, the main recommendations are that the service should be conducted in the band $88-95 \mathrm{Mc} / \mathrm{s}$ and that a start should be made on a fairly large scale. The Committee points out the radio industry should be given encouragement to embark on the production of suitable receivers in reasonable volume. This recommendation, if accepted, is a matter for some satisfaction, and is in welcome contrast to the Government's proposals for starting competitive television on a severely restricted basis.

It should be added that the Committee's recommendations apply either to a scheme designed to give national coverage of three programmes or to a network of stations planned to give merely "a substantial reinforcement " of the services at present provided by the B.B.C. on long and medium waves,

A lengthy report from the Technical Sub-Committee is appended to the document, and in it the relative merits of frequency and amplitude modulation are discussed from the point of view both oi transmission and reception. As the main point at

[^0]issue has now been finally decided by the P.M.G.'s acceptance of the Committee's recommendation, there is little value in commenting on these arguments in detail. Considerable stress is laid on the greater capital and upkeep cost of a.m. transmitters, and also on the economics of receivers for the two modulation systems. The conclusions reached are, however, simple and straightforward: "The a.m. wide-band system has specific disadvantages and no advantages relative to the f.m. system. For the a.m. narrow-band system the only advantage is a possible saving in the frequency spectrum required."

Perhaps the most interesting part of the Technical Sub-Committee's report, now that the a.m./ f.m. issue is settled, is that dealing with some of the details of f.m. receiver and convertor design. We are pleased to see a recommendation (which is endorsed by B.R.E.M.A.) that standards of good practice should be set up from the start in order to minimize harmful radiation from v.h.f. broadcast receivers. These problems should be studied jointly by the Post Office and Industry.

In general, the discussion on f.m. is based on the acceptance of the almost universal standards of a maximum deviation of $\pm 75 \mathrm{kc} / \mathrm{s}$ and a modulation bandwidth of $15 \mathrm{kc} / \mathrm{s}$.

Just as the Report opens on a rather doleful note, it ends in similar vein. There is a minority statement signed by C. O. Stanley, who considers that adequate attention has not been given by his colleagues to the broader questions of fundamental policy. Mr. Stanley says v.h.f. broadcasting has been a failure in practically every country in which it has been introduced, and casts strong doubts on its advantages in general. He also makes a point of the "incompatibility" of f.m., as recommended for the new service, with a.m. as used for all existing broadcast sound, and which, presumably, will be used for the proposed competitive television service. This last is a provocative point, and in this matter at least Mr. Stanley will probably find some supporters among those who believe that television and sound broadcasting may ultimately be merged into a more-or-less integrated service.

## 

The following appreciation of Major Armstrong, whose tragic death is recorded on page 124, has been received from Capt. H. J. Round, who has been responsible for many radio developments in parallel fields.

The tragic death of Armstrong has given a sincere shock to all his friends, many of whom are on this side of the Atlantic.

The writer first met him in 1917 during the later stages of the war when he was a Major in the U.S. forces then going into Europe, and the friendship then established has been maintained throughout the years.

I had the good fortune to be given a very early demonstration of his superheterodyne in the Paris laboratory he had established, and, as the world knows, this basic invention was followed first by super-regeneration and in 1935 by wide-band frequency modulation.

This, as with some of his other inventions, involved him in very prolonged and expensive litigation, a good deal of which his closest acquaintances thought could have been avoided. However, the intensity with which he attacked technical problems he also applied to his legal problems and there is not much doubt in my mind that this double load clouded his very great intellect in the end.

Since those carly years I have met him a few times and have been in constant correspondence with him, and I am happy to think that only a year ago I was able to spend considerable time with him in New York. A sentence of Armstrong's own in his recent paper on "The Spirit of Discovery," in which he culogizes the work of Marconi, I think applies very
well to Armstrong himself: "It is seldom that a man makes two basic discoveries. When a man makes three, his attitude towards problems and his method of work merit close analysis and study." Armstrong should go down in American history as one of her great sons, worthy to be classed with Edison, Bell and Westinghouse.

## MNTERFEREXUE LIMITS

THE publication by the British Standards Institution of a revision of the Standard specifying the limits of radio interference is of particular interest in view of the announcement that the two committees appointed to advise the P.M.G. on interference from small electric motors and refrigerators have now submitted their reports and, too, that he hopes shortly to lay regulations before Parliament.

The revised BS 800:1954* now covers Band 1 in addition to the long- and medium-wave bands. The limits of magnitude of noise voltages measured from each line terminal to earth are laid down as $750 \mu \mathrm{~V}$ ( $40-70 \mathrm{Mc} / \mathrm{s}$ ) and $1,500 \mu \mathrm{~V}(200-1,605 \mathrm{kc} / \mathrm{s})$ and the limits of noise fields at a distance of 10 metres are $50 \mu \mathrm{~V} / \mathrm{m}$ and $100 \mu \mathrm{~V} / \mathrm{m}$ for the respective frequency ranges. Measuring apparatus and methods of measuring are specified in BS 727:1954, $\dagger$ which has been revised and now covers the range $150 \mathrm{kc} / \mathrm{s}$ to $150 \mathrm{Mc} / \mathrm{s}$.

* " Limits of Radio Interference," price 4 s.
+ " Characteristics and Performance of Apparatus for Measurement of Radio Interference," prize 4s.


## L.MII IVTEREEREXCE

THERE was some mild controversy in our correspondence columns last year over the question of radio interference from electric lamps. The writers of most of the letters contended that radiation was restricted to vacuum lamps, but others stated that trouble was also caused by gas-filled types.

The facts seem to be that neither kind is entirely free of blame, though the vacuum lamp is by far the most serious source of trouble, as it can radiate interference throughout its life and when in apparently perfect condition. Radiation from the gas-filled lamp,
on the other hand, usually occurs when the bulb is coming to the end of its life, and is produced by arcing across a minute break in the filament.

So far as television receivers are concerned, interference from vacuum lamps is generally of the nature shown in the accompanying illustrations (by John Cura). The double line of interference, as in the middle picture, is perhaps the most typical, though the single line on the left is not uncommon. The right-hand picture is an enlargement which shows (at top) the broken nature of a typical interference line.

Characteristic interference patterns produced on a television receiver by a nearby 60 -watt vacuum (traction type) lamp.


#  <br> Squandered V.H.f. Channels 

## I'hy There is a Lack of Space for T'elerision and Mobile

Radio Derelopments

By J. R. BRINKLEY*

T has been suggested in these columns and elsewhere that the administration of radio frequencies in this country is not working satisfactorily. The situation which has come to light as a result of the pressure for frequencies in Band 3 for additional television services is certainly disturbing and this article has been written in an attempt to clarify as far as possible in a short article what has become a very complicared situation.
The block allocation of the very high frequencies in the United Kingdom is shown at the head of this article and, in conjunction with the diagram, the following three points should be noted: -

1. The "single programme" service of the B.B.C. operating in the band $40-68 \mathrm{Mc} / \mathrm{s}$ has room for five 405 -line TV channels. This block is probably more than adequate for nation-wide coverage with one programme.
2. The desire for competitive televisior in this country no doubt stems greatly from the example to be found in the U.S.A. where most large cities have a choice of two or three programmes and where New York has six and Los Angeles seven programmes, all operating on the immediately practicable frequencies below $216 \mathrm{Mc} / \mathrm{s}$.
3. If the total frequency spectrum available below $216 \mathrm{Mc} / \mathrm{s}$ to sound and TV broadcasting in the U.S.A. (a total bandwith of $92 \mathrm{Mc} / \mathrm{s}$ ) were currently available in Britain, there would be no difficulty in providing multi-programme TV. It would, in fact, be rather easier, since a British TV channel requires only $5 \mathrm{Mc} / \mathrm{s}$ as against the American $6 \mathrm{Mc} / \mathrm{s}$.

Why is it that instead of a total of $92 \mathrm{Mc} / \mathrm{s}$, as in the United States, probably something less than
$10 \mathrm{Mc} / \mathrm{s}$ can be found in this country fo: the immediate expansion of television? The answer is, that unlike the U.S.A. and most of the rest of Europe, the U.K. did not reserve adequate frequencies for television at the last International Conference (Atlantic City, 1947) and that subsequently, some of the frequencies which were reserved for television have been given to or taken by other services. Under the Atlantic City agreement, the world is divided into three regions and the U.K. and Europe are in Region 1. The Region 1 allocations to broadcasting (below $216 \mathrm{Mc} / \mathrm{s}$ ) are $41-68 \mathrm{Mc} / \mathrm{s}, 87.5-100 \mathrm{Mc} / \mathrm{s}$ and 174 $216 \mathrm{Mc} / \mathrm{s}$-a total of $81.5 \mathrm{Mc} / \mathrm{s}$. If these bands were available in the U.K., they would be adequate for immediate broadcasting requirements.

Of these three bands, however, it is now apparent that only the "B.B.C." block (Band 1) is available intact. There are two reasons for this: first, Atlantic City footnotes inserted by the British delegation which allocated parts of them to other services, and second, subsequent "national" allocations made in this country to other services. As a result, frequencies have been "lost" on the $87.5-100 \mathrm{Mc} / \mathrm{s}$ band to the police and fire services ( $95-100 \mathrm{Mc} / \mathrm{s}$ ). Further frequencies have been "lost" in the $174-216 \mathrm{Mc} / \mathrm{s}$ to mobile and "fixed" services (174-184 Mc/s) and, in the band $200-216 \mathrm{Mc} / \mathrm{s}$, to aeronautical navigation services (Distance Measuring Equipment-DME). Since the band $87.5-95 \mathrm{Mc} / \mathrm{s}$ is intended for sound v.h.f. broadcasting, the outcome of all this is that the only frequencies at present available for TV expansion lie between 184 and $200 \mathrm{Mc} / \mathrm{s}$. But it is important to

[^1]note that when guard bands and other factors are considered, even this small band is probably not available intact.

We shall now examine the "intruding" services to see how they got where they are.

Aeronautical Navigation Services-200-216 Mc/s The authority for operating these services is drawn from footnote 89 to the Atlantic City agreement which states-" In the U.K. DME will be operated on the band $200-235 \mathrm{Mc} / \mathrm{s}$ until such time as world standardization (of DME) on $1,000 \mathrm{Mc} / \mathrm{s}$ has been accomplished." This footnote precludes the use of TV in this band since it is obviously incompatible with DME. On the other hand, continental Europe plans TV stations in this band authorized by the agreement, which would make $200-\mathrm{Mc} / \mathrm{s}$ DME unworkable in the U.K. It will be observed in passing that world standardization of DME on $1,000 \mathrm{Mc} / \mathrm{s}$ was an even more remote possibility in 1947 than it is at this moment.
It is unfortunate, therefore, that these frequencies were earmarked for a service which is prejudiced by Continental TV planning, and yet at the same time could only be used by British TV stations by closing down such services and ignoring the Atlantic City provisions. (Footnote 87 states categorically-" the band $200-216 \mathrm{Mc} / \mathrm{s}$ is allocated for the aeronautical radio navigation service.")

Fixed Services-174-200 Mc/s.-These are point-to-point G.P.O.-type telephone links (e.g., between islands in the Channel Isles). International authority for them is obtained under footnote No. 87 which states "In the U.K., the band $174-200 \mathrm{Mc} / \mathrm{s}$ is also allocated for the fixed service." This footnote was not very realistic since with anything but exclusive band allocations, the planning of broadcasting services will prove unnecessarily difficult.

Mobile Services-174-184 Mc/s.-The occupiers of these frequencies include mobile services operated by


Fig. 1. This diagram shows the increase of wanted signal level required to restore the signol-to-noise ratio after the Alexandra Palace TV transmitter has been switched on. The out-of-band radiation was measured at $1 \frac{1}{2}$ miles from the television station.
ambulances, railways, the Press, industrial users, electricity, gas and oil undertakings and the majority of radio taxi services. These services have been operating in this band since 1948-9. Such services were first allocated frequencies in the $70-\mathrm{Mc} / \mathrm{s}$ band, but in 1948 the G.P.O. made it known to manufacturers that all future allocations would be in the $156-184-\mathrm{Mc} / \mathrm{s}$ band which would be the "permanent and secure band for such services." This was a great embarrassment at the time since no suitable apparatus or valves were available for these frequencies. After 12 months development, fitting in the new band began and has been proceeding ever since. It was not realized, and the G.P.O. did not point out, that this allocation was not in accordance with Atlantic City. The recent official statement by the PostmasterGeneral that these frequencies will be cleared for TV has created a situation in which manufacturers and users, who have invested heavily in this band, have lost confidence in all official statements regarding the allocation of mobile frequencies. Their confidence will only be restored by adequate compensation for loss of frequencies and capital investments and written Government undertakings giving long-term security in any new arrangements made for them.

One of the outstanding shortcomings of the Atlantic City agreement and most of the subsequent planning provisions is that virtually no account is taken of guard bands. Thus, for example, in continental Europe one country may operate mobile stations up to the limits of the International allocation; for example, up to $174 \mathrm{Mc} / \mathrm{s}$, and an adjacent country may operate TV down to $174 \mathrm{Mc} / \mathrm{s}$. In the border areas, the two types of transmission will experience serious mutual interference, but both countries will be operating within their rights. A very substantial guard band will be required to obviate such interference, but no agreed assessment exists as to the extent of such guard bands or as to how their provision will be shared between the adjacent services.

This problem is equally important within the borders of any one country. In the U.K. it has a most significant bearing on the present Band 3 problems. The interference suffered will be two-way. It will be caused to the mobile service because as can be seen from Fig. 1, TV transmitters radiate appreciable power several megacycles beyond their allocated channels. Likewise, interference will be caused to the TV service because of the very poor selectivity employed in most TV receivers. An example of sound


Fig. 2. Comparisons of the selectivity of different receivers. A, sound channel of typical television receiver; B, German domestic v.h.f. receiver; $C$, typical mobile v.h.f. receiver.


Fig. 3. Estimated interference from mobile services to TV receivers. The curves show the input of interference needed at the television receiver and frequency separation from the vision carrier for various conditions of interference. Curves $A(300 \mu \mathrm{~V})$ and $B(30 \mu V)$ are for interference just sufficient to cause picture tearing, while $C\left(30 \mu_{\mu} V\right)$ and $D(30 \mu V)$ are for just discernible interference. Curves $E\left(170_{\mu} V\right)$ and $F\left(25_{\mu} V\right)$ are for just audible interference on the sound channel. The curves are for the TV signal levels shown in brackets. Scoles are added to relate the interference input to distance in yards between the TV aerial and the aerial of a mobile transmitter (1) and a fixed transmitter (2), in both coses for a power of ISW.
channel selectivity characteristics is shown on Fig. 2. Television receivers have so far been designed without any serious attention being paid to providing adjacent-channel selectivity, and the effect of this is shown in Fig. 3, in which interference to a TV receiver from a mobile transmitter is estimated for various ranges.
This article is not necessarily a plea for adopting American methods of frequency allocation which have their own shortcomings and are being currently criticised. It is to be noted, however, that the present U.S. methods have made multi-programme TV widely available in that country, while "frequency shortage" is making it more than difficult in this country. Direct comparisons are not easy, but the U.S. achievement does not appear to have been at the expense of mobile radio development. There are estimated to be more than 300,000 radio equipped mobiles operating in the U.S. as against fewer than 10,000 in the U.K.

In America, the Federal Communications Commission has been able to publish a great deal of documentation on the subject of frequencies, much of which is well conceived. It is specific in matters of policy and detail in a way in which it is of the greatest possible value to manufacturers and operators alike. No such documentation is available in this country.

It is suggested that three preliminary steps are necessary to put frequency allocation on a sound basis. First, there must be a clear legal basis for frequency allocation. The only existing legislation relating to frequencies is the Wireless Telegraphy Act. This gives the P.M.G. the widest possible powers relating to the establishment of transmitting stations. But it has been stated that the P.M.G. does not regard himself as responsible for "overall" frequency allocation which is carried out by an "Inter-departmental Committee." This vital function is, therefore, apparently carried out by an anonymous committee for which
the writer can discover no legislative authority. Such an arrangement is not likely, in the writer's opinion, to produce the best results.
Secondly, there must be a clear declared policy in matters relating to frequency allocation. As an initial step in establishing a policy, the Government must make up its mind whether or not it intends to stand by the international frequency agreements to which it is signatory.
Thirdly, there must be adequate independent machinery to administer the policy laid down in an equitable manner. The writer suggests that the American F.C.C. is by far the most adyanced example of such machinery to be found anywhere in the world to-day.

## - Introduction to Valvas.

MOST radio textbooks either take it for granted that the reader knows all about valves or deal with the subject rather sketchily. On the other hand, books written about valves specifically are often too highbrow and specialized, not to mention expensive. Where, then, is the radio man with only a nodding acquaintance of the subject to get the sort of information he wants? The answer is to be found in "Introduction to Valves" by R. W. Hallows, M.A.Cantab., M.I.E.E., and H. K. Milward, B.Sc.Lond., A.M.I.E.E. This is a comparatively small book, but a careful selection of material has ensured that its 152 pages and 107 illustrations provide just the right kind of information for everyday radio work.

An important feature of the book is that it explains the operation of the valves in typical circuits-they are not just left in mid-air, so to speak. Beginning with an exposition of fundamental principles, there are chapters on diodes as rectifiers and detectors; triodes; tetrodes and pentodes; multiple-grid valves for frequency changing; power-output valves; and valves for v.h.f. and c.h.f.

The book can be obtained from any bookseller, price 8 s 6 d , or direct from our publishers at 9 s by post.

# Cathode Follower Oscillator 

L'sing $R C$ Netzoorks with a Voltage Step-up

By THONAS RODIDAM

IN an article in last month's issue I described an oscillator circuit in which the valve was connected as a cathode follower. The valve in this circuit really uses a cathode follower, too, with a sufficiently high cathode load for the negative feedback to dominate the system completely. The circuit was analysed as a particular circuit, but it was mentioned that it could be regarded as a member of a class of oscillators, those using over-balanced rejector circuits. In fact I had overlooked one variant, which is not a rejector circuit at all, so that the classification becomes just that of a cathode-follower oscillator. Most of these use only resistances and capacitances (or inductances), although last month's variant included all three types of element, and it is of some interest to survey the properties of the group as a whole.

The feature of the cathode-follower oscillators which makes them especially interesting is that most of them use a resistance-capacitance network to give a voltage step-up. When we think of an RC network we always see ourselves carrying out a lot of voltage divider calculations, and getting out less than we put


Fig. 1. The basic feedback amplifier or oscillator consists of the amplifier $A$ and the feedbock $\beta$.

Fig. 2. A passive quadripole with two input terminals, 1,2 and two output terminols, 3,4.


Fig. 3. If terminals 2 and 4 have an internal connection in Fig. 2, the system can be drown like this. The input terminals are still 1,2, but the output can be taken from 3,2 or 3,l.
in : it isn't necessarily so, however, and the way in which these circuits provide the step-up is so simple that it seems incredible that no one should have spotted it before 1947.* We need a step-up, of course, if we want to use a cathode follower, because we can never get quite as much at the cathode as we put in at the grid. The step-up need not be very large.

First of all, let us look at the basic feedback oscillator. This is one way of looking at the general idea of an oscillator, a way which happens to be very convenient for purposes of analysis. The circuit is shown in Fig. 1 and it looks, indeed is, just our old friend the feedback amplifier. As you will remember, the basic equations for this take the form :

$$
\begin{gathered}
V_{2}-A\left(V_{1}-V_{3}\right) \\
V_{3} V_{2} \beta=A V_{1} \\
\text { so } V_{2}(1-A \beta)=A \\
\frac{V_{2}}{V_{1}}=1+A \overline{A B}
\end{gathered}
$$

These are the equations in the form most convenient for amplifier working, with no minus signs to think about: some people prefer to put in a minus sign in the appropriate place, and with the gain as $m=$ $\mu(1-\mu \beta)$. It boils down to the same thing in the end.

As you can see, if $\mathrm{A} \beta=-1$, the gain will be infinite, so that if we close the input terminals through a finite but very small resistance, the noise in this resistance will still be amplified up to the overload point of the circuit. In fact, it will oscillate, and that bit about noise was just to show how the oscillation starts. A practical oscillator will always have $A \beta=$ $-k$, where $k>1$, with very low level conditions, and the amplitude of oscillation will grow until something in the circuit alters $A$ or $\beta$ to reduce $A \beta$ to unity. It is perfectly possible for such a circuit, with $-A \beta>1$, to be stable, and amplifiers which are " conditionally stable " have in fact been used. One of the advantages of the feedback amplifier approach to oscillator theory is that it helps to clarify the behaviour in these special cases, which are often associated with " mode jumping " and other awkwardnesses.

The other way of treating oscillators is to lump some components with the valve into a package which gives a negative resistance across two terminals. Transistors, with their built-in positive feedback, are easily considered in this way. Again we have two boxes, but the way the elements are split between them is different. The final answer must, of course, be exactly the same.

It will be clear that the problem which confronts us in designing a cathode-follower oscillator is that of obtaining a suitable network for the $\beta$ box, remember-

[^2]

Fig. 4. Resistancecapacitance circuit used in the phaseshift oscillator.

Fig. 5. Vector diagram for the voltages in Fig. 4.

ing that in the box marked A we have a gain of about $0.9-0.95$ and no phase shift.

Let us consider the "passive quadripole" shown in Fig. 2. Just a box, with four terminals, of which two are called inputs ( 1,2 ) and two called output ( 3,4 ), looking unpleasantly like the introduction to some heavy mathematics. We can simplify matters by joining 2 and 4 together, because all the systems we want to discuss actually have got an internal connection here, which we usually think of as earth. This is the trap into which we must not fall. Let us re-draw the circuit in a more non-committal form (Fig. 3). The voltage arrows now run clockwise between each pair of terminals, and a new one, $\mathrm{V}_{3 \cdot 1}$, has been dotted in. When we apply a voltage $V_{12}$ across the input terminals 1,2 , we can now say that we get an output of $V_{3: 1}$ across 3, 4 in Fig. 2, or outputs of $V_{2: 3}$ ( $=-V_{3.4}$ of Fig. 2) across 2, 3 in Fig. 3 or an output of $\mathrm{V}_{3: 1}$ across 3, 1 in Fig. 3. Quite clearly we must have $V_{3 \cdot 1}=-\left(V_{1 \cdot 2}+V_{2 \cdot 3}\right)$, and looking at Fig. 2 again this means $V_{1: 3}=$ $V_{1 \cdot 2}-V_{1: 9}$.
Suppose that the network inside the box in Fig. 2 produces a loss of $n$ times, with a phase shift of 180 degrees. Then we shall have $\mathrm{V}_{3 \cdot 4}=-n \mathrm{~V}_{1.2,2}$, when $n$, of course, is a proper fraction. Obviously $\mathrm{V}_{1,3}=$ $(1+n) V_{1 \cdot 2}$. We can con-- sider terminals 1, 2 as input, and 1,3 as output, and we have got a step-up of $1:(1+n)$ in this network, which we thought was just attenuating and shifting phase.

So far we have been discussing the rather depressing "passive quadripole" but it may make things rather clearer if we consider a specific circuit. The circuit shown in Fig. 4 is the one we associate with the phase-shift oscillator. The limiting case of this network is when $C_{1} R_{1}=C_{2} R_{2}$, $=C_{3} R_{3}$ and $R_{1} \ll R_{2} \stackrel{2}{2}$ $\mathrm{R}_{3}$, and this limiting form
is particularly easy to analyse. At one Erequency each RC section produces 60 degrees phase shift, so that the vector diagram shown in Fig. 5 is obtained, with $V_{A}$, the voltage across $R_{1}$, equal to one-half $V_{1 \cdot 2}$ and rotated 60 degrees. The following RC circuits each divide the voltage by 2 , giving $\mathrm{V}_{3,4}=-\frac{1}{4} \mathrm{~V}_{1,2}$.

As we have already shown, $V_{1 \cdot 3}=(1+n) V_{1: 3}$, where $n$ is here equal to $\frac{1}{x}$. Redrawing the circuit in the way shown in Fig. 6(a) we have a rather odd-looking network which provides a step-up ratio of $9: 8$ although it uses only resistance and capacitance elements. The network shown in Fig. $6(\mathrm{~b})$ is the step-up version of the other phase-shift oscillator circuit, the one with the C's and R's interchanged, but this time it has been arranged rather differently to bring out two features of the circuit, the common connection at one side of the capacitances and the d.c. path on the output side.

These two circuits are often constructed with equal values of C's and R's, and the output voltage from the basic network is then only 129 times the input. This means that when twisted round we get only $30 / 29$ times the input, which is very tight indeed for use with a cathode follower. There are various compromise solutions, all of which are inconvenient in one way or another.

## Choice of Circuit

If we want to make the collection of diagrams more impressive, we can consider the 4 -stage RC networks and also the corresponding 3 - and 4 -stage RL networks. There is a great deal of tedious algebra already published about these networks. Some of it can be replaced by common sense : for example, it needs no mathematics to see that the arrangement of Fig. 6(b) lets harmonics through easily, while if it were an RL network the harmonics would be attenuated. It needs no algebra to show that the input impedance of the Fig. 6(b) network is more than $R_{1}$, and in the


Fig. 6. The circuit of fig. 4 can be redrawn either as in (a) or, if we take the output across 3,1, as in (b).
Fig. 7. The parallel-T on the left transforms to that on the right in its step-up form.

limiting graded case is $2 \mathrm{R}_{1}\left(\mathrm{R}_{1}+j \sqrt{ } 3 \mathrm{R}_{1}\right)$. In a practical design this input impedance will be part of the cathode load of the cathode follower, and the value of $R_{1}$ must, therefore, be high enough to provide the proper gain conditions.

The networks we have been considering so far have been phase-shift networks with some attenuation, but the attenuation has been of the smoothly falling kind : monotonic, the mathematicians call it. The only condition is that the network must be complicated enough to give more than 180 degrees phase shift.

We must now turn to the other sort of RC network, of which the parallel-T is the most familiar example. This network, in its basic form and its twisted round form, is shown in Fig. 7. We could write down the network equations and find out what happens: the answer, in the particular case, when $\mathrm{R}_{2}{ }^{\prime} \mathrm{R}_{1}=\mathrm{C}_{3} / \mathrm{C}_{2}=$ $m \rightarrow \infty$, and $\mathrm{R}_{3} / \mathrm{R}_{\mathrm{t}}-\mathrm{C}_{3} / \mathrm{C}_{1}=n=0.41$, gives a step-up of 1.207 under zero phase conditions. A simpler way of secing what happens is to draw the network in the form shown in Fig. 8 and build up a vector diagram on the assumption that the $\mathrm{R}_{2} \mathrm{C}_{2}$ arm does not load the rest of the network does not load the rest of the network


Fig. 8. Another way of drawing the transformed parallel-T circuit, and (right), a vector diagram of the network voltages.


Fig. 9. Four-phase step-up network giving a meosured goin of 1.67 ot $640 \mathrm{cs} . R_{1}=100 \mathrm{~s} 2, R_{2}=1 \mathrm{k} \mathrm{\Omega}, R_{3}=50 \mathrm{ks}$, $C_{1}=2.5 \mu F, C_{2}=0.25 \mu F, C_{3}=0.00 \tilde{S}_{\mu} F$.
 the algebraists have done in taking $m \rightarrow \infty$. The resulting vector-diagram is shown in Fig. 9, and gives a clear indication of how the voltage step-up occurs.

You can have a busy time extending Fig. 8 too, and Fig. 9 (IWireless Engineer Jan. 1953, Fig. 6, p. 21 ) shows how Bacon and Salmon have done this. Their network, with the values shown, was calculated to give a stepup of 1.98 times, and actually gave 1.68 times at 640 c s .

Just as the parallel-T network gives, in its twisted-round form, a desirable step-up, so does the series- $\pi$, its dual. As you would expect, the series- $\pi$ gives a current step-up, but by working it back to front the wanted voltage step-up is obtained. A large collection


Fig. 10. The series-m network on the left can be twisted round to give the step-up network on the right. This form is especially convenient for use with valve circuits.

These oscillators which use the balanced type of network must not be confused with the parallel-T oscillators which are already well known. The common form of the parallel-T network applies negative feedback through the parallel-T, and some positive feedback through a separate resistive network. The negative feedback keeps theamplifierfrom oscillating except at the frequency at which the network is balanced. Here the only feedback is through the network, which does not operate at the normal balance point, but at what may be termed a special over-balanced point.

There is not much more to be said about these oscillator circuits, unless we settle down to calculate particular values. A great deal of the analysis has already been published by Dunn, and anyone who wishes to study the matter in more detail is recommended to refer to his paper and to the references


Fig. 11. (a) Basic oscillator derived from the circuit of Fig. 10, together with, (b) a modification, $R_{b}$ and $R_{a}$ replacing $R$, to get suitable bios conditions.
given at the end of it. This paper also discusses the use of the system under non-oscillatory conditions, as a selective amplifier. Feedback amplifiers incorporating RC networks are extremely useful for very low frequencies, where high values of $Q$ cannot be obtained easily with LC networks, but this is outside the scope of this article.

## CHICIB

Birmingham.-The March 5th meeting of the Slade Radio Society will be held at the Imperial Hotel, Temple Street, Birmingham, at 7.30 , when a film-illustrated talk on valve manufacture will be given by a representative of Mullard's. Sec.: C. N. Smart, 110, Woolmore Ruad, Erdington, Birmingham, 23.
Brighton.-Meetings of the Brighton and District Radio Club continue to be beld at the Eagle Inn, Gloucester Road. Brighton, every Tuesday at 7.30. The club transmitter (G3EVE) is on the arr on the second Tuesday of each month on 80 metres using both 'phone and c.w. Sec.: T. J. Huggett, 15, Waverley Crescent, Brighton.

Cleckheaton. -"Crys?al Microphones and Pickups" is the subject of the talk by N. G Newman, of Rothermel, Lid., to be given at the mecting of the Spen Valley and District Radio and Television Society on March 24th. Mectings are held on alternate Wednesdavs at 7.30 at the club headquarters in the Temperance Hall, Cleckheaton. Sec.: N. Pride, 100, Raikes Lane, Birstall. Nr. Leeds.
Romford.-"Television Interference and Transmitter Design" is the subject of the talk to be given by Louis Varney, A.M.I.E.E. (G5RV). to members of the Romford and District Ar oteur Radio Society on March 30th. Meetings are held each l'uesday at 8.15 ai R.A.F.A. House, 18, Carlton Road, Romford. Sec.: N. Miller, 10, Rom Crescent, Romford.
S.E. London.-Meetings of the Clifton Amateur Radio Society (G3GHN) are held every Fridsy evening at the society's headquarters, 225, New Cross Road, S.E.I4. Sec.: C. H. Bullivant (G3DIC), 25, St. Fillans Road, Catford, S.E.6.

Southend.-At the meeting of the Southend and District Radio Society on March 19th, G. T. Peck, of Ernest 'I'urners, High Wycombe, will speak on the radio control of models. The society meets at the Municipal College Laboratories, Queens Road, Southend-on-Sea, on alternate Fridays. Morse and theory classes are again being held at 27. Park Road, on l'uesdays and Thursdays at 8.0. Sec.: J. H. Barrance, 49, Swanage Road, Southerd-on-Sea.

Wellingborough.-A talk on television acrials will be given by J. W. Hoblev at the March 18th meeting of the W'ellingborough and District Radio and Television Society. The club meets each Thursday at 7.30 at the C.W.S., Silver Street, Wellingborough. Sec.: R. J. Henty, 6B, Silver Strees, Wellingborough.

## NEWN

Wolverhampton . smateur Radio Society recently moved to new headquarters at Stockwell End, Tettenhall, where the club transmitter (G85A) is being installed. The club meets fortnighty on Mondays with morse classes on the alternate Mondays. Sec.: H. Porter (G2YM), Applegarth, 221, Park L.ane, W'olverhampton.

QRP.-Readers interested in the low-power operation of transmitters, whether it be for communication or the control of models, are invited bv the QRP Research Socicty to send for details of membership and a specimen copy of the Society's iournal. Contests for both transmitters and short-wave listeners are held throughout the year and a QRP ne: is organized each Sunday at 2.30 on $1.9 \mathrm{Mc} / \mathrm{s}$. Sec.: J. Whitehead, The Retreat, 92, Rydens Avenue. Walton-on-Thames, Surrey.
Radio Control.-At the meeting of the Birmingham group of the International Radio Controlled Models Society on March 6th at 2.30 at the International Centre, Suffolk Street, Birmingham, J. Merrick will speak on "A Reliable Receiver," A demonstration of a radio-controlled model land vehicle will be given by R. F. Stock at the meeting of the London group on March 1 tith at 2.0 at the Horseshoe Hotel, Tottenham Court Road, London, W.1. Sec.: C. H. Lindsey, 55, Tenison Road, Cambridge.

## Standard Valve Symbols;

ADDITIONAL symbols for electronic tubes and valves, including gas switches, are given in Supplement No. 3 (1953) to B. S. $530: 1948$ (Graphical Symbols for Telecommunications). The basic electrode symbols are given, together with examples of their assembly to represent complete valves. The emphasis is on cold-cathode discharge valves and many of the symbols are unfamiliar ones. Travelling-wave valves, cavity magnetrons, velocitymodulation valves and photo-voltaic cells are included, as well as TR ceils or gas switches. The Supplement is issued by the British Standards Institution, 2, Park St., London, W.I, price 2s 6 d .

# Measurement of Harmonic Distortion <br> <br> Self-contained Direct-reading Instrument for <br> <br> Self-contained Direct-reading Instrument for <br> Works Testing and Servicing 

By 'I'. I). (i()NW AI, ${ }^{\star}$ B.s.s.(Eng.). A.C.(i.I., A.M.I.E.E.

NUMEROUS articles have been written on the general aspects of audio distortion, its detection, and cure, but these notes are intended to cover the specifically practical problem of measuring distortion on a mass production flow-line, and objectively assessing distortion in a service department handling audio equipments. Many methods of distortion measurements are already available, but on examination they were all found to require skill and time to give accurate results, whereas what was wanted was a direct-reading instrument capable of giving accurate repeatable measurements with unskilled labour.

This type of measurement is becoming of great importance in the manufacture of tape recorders, where it is usual to set the maximum recording level to correspond to a definite amount of distortion, which is a compromise between dynamic range and quality. The increasing emphasis on fidelity in amplifiers makes it necessary to check quickly performance figures of output against distortion to verify that the equipment meets the published specification.

In this article it is proposed to deal only with harmonic distortion, that is to say harmonics produced in the output of a system when a single frequency is applied to it : normally the amount of harmonics produced will depend upon the input signal and hence it may often be necessary to plot a curve relating input to distortion in the output. The distortion factor of a periodic voltage is the ratio of the total r.m.s. voltage of the harmonics (i.e., the square root of the sum of their squares) to the total r.m.s. voltage. When this figure is multiplied by 100 it gives what is known as the percentage of total harmonic distortion, or more simply percentage of total distortion.

In some cases where we know that a certain harmonic is more pronounced than others we may consider only its ratio to the total r.m.s. voltage, and this we will call the percentage of second or third, etc., harmonic distortion. In tape recorders the third

[^3]harmonic is dominant and of chief interest, as it is in push-pull amplifiers. In Class A triodes the second harmonic is dominant, and in pentodes all harmonics up to the fifth are usually significant.

When the percentage of total distortion is small (i.e., less than 10 per cent), it may be more simply expressed as the square root of the sum of the squares of the individual harmonic percentages. The calculated error of such a simplification is less than one part in two hundred,
percentage 2 nd harmonic, $k_{3}=$ percentage 3 rd harmonic, etc.
In order to make a comprehensive assessment of distortion on a piece of equipment, measurements should be carried out over the whole of the working frequency range, due regard being paid to neglect those harmonics which lie outside the audible range. The test procedure is lengthy and considerable skill is needed to give the correct results. A simple system should be usable by unskilled labour, involve no calculation or charts, and preferably be direct reading.

Since we are more concerned here with the testing of distortion, the first simplification to suggest itself is the use of a single frequency, and to provide that frequency with the distortion-measuring circuit in a

Fig. I Colpitts oscillator and cathode follower with less than 0.25 per cent distortion.



In the Grundig Type TGS distortion meter, controls associated with the 1,00J-c s oscillator are on the left and those for range selection and calibration on the right.
from the original signal the phase relationship on any equipment should be constant.

An equipment was developed on the above lines, provision being made to measure up to 10 per cent of third or total harmonic distortion by direct reading, for a $1,000-\mathrm{c}$ s oscillator of low harmonic content having an output continuously variable up to 10 volts, and for the accommodation of inputs between 10 millivolts and 100 volts.
Oscillator Unit. The oscillator needs to be stable in operation and to have as
single self-contained unit. The next problem is to decide which individual harmonics are of interest, or alternatively whether the total harmonic distortion is to be measured. For our particular problems a frequency of $1,000 \mathrm{c}$ 's was chosen, and it was decided to provide facilities for measuring either the percentage of total distortion of that of third harmonic distortion.

As explained previously a true measurement of total distortion involves the r.m.s. summation of all harmonics, and this can only be carried out by using either a thermal instrument, square-law valve voltmeter or a suitable dynamometer. For practical purposes these all have to be ruled out, and it was decided to carry out some tests on a " 1 -milliamp" full-wave bridge instrument rectifier, in conjunction with a 100 -microamp meter. First, it was found that the superimposition of 10 per cent of second or third harmonic on the fundamental only affected the reading by less than $\pm 1$ per cent as its phase was changed, hence as an indicator of the r.m.s. value of a distorted wave it is sufficiently accurate. Secondly, 5 volts of $2,000 \mathrm{cs}$ and 5 volts of $3,000 \mathrm{c} / \mathrm{s}$ fed together gave a deflection of 6.8 volts, which is within reasonable limits of the calculated value of 7.07 ; again phase variation between the two can cause a difference, but since all harmonics will be generated
low a harmonic content as possible. A Colpits circuit was chosen and arranged as an electroncoupled oscillator to buffer the tuned circuit, with a tuned anode to reduce the output of harmonics. In order to maintain the minimum of harmonic content the amplitude of such an oscillator must be closely controlled, and as a first approach the h.t. supply was regulated by a neon type stabilizer. Next some attempts were made to limit the amplitude by negative feedback, but although this produced excellent results, satisfactory starting could not be guaranteed.

Finally a germanium diode was used between the anode of the oscillator and the cathode via a variabl: feed resistor which gave perfect control of amplitude. When oscillation starts the diode acts as a hig' impedance across the tuned circuit, biased by the cathode-to-earth voltage, and does not conduct. As the amplitude of oscillation increases the diode hegins to conduct and increases the cathode potential of the oscillator, thus automatically controlling its amplitude. A high-impedance d.c. return for this current is provided by a second diode.

The final arrangement as shown in Fig. 1 was extremely stable and had a harmonic content of less than : per cent. A cathode follower was used for the output stage to give complete isolation and to provide a simple form of output attenuator. Since

Fig. 2. A three-section filter, with toroidally wound inductonces, is used to suppress the fundamental.



Underside of chassis showing toroidal filter coils under RC sub-assemblies.
a trace of second harmonic was introduced by the cathode follower, a simple acceptor circuit was added across the cathode load, bringing the total distortion content of the output down to less than $\not \pm$ per cent.

Distortion Measuring Unit. The design of filter units is greatly simplified by the use of lowimpedance values, and this in turn limits the dangers of stray pick-up and removes the necessity for shielding. For the inductances toroidal windings were used, since high Q's can be readily obtained and the stray field is virtually negligible.

To remove the fundamental $1,000 \mathrm{c} / \mathrm{s}$ a three-section $m$-derived high-pass filter was employed, fed from a cathode follower input of approximately 200 ohms output impedance. Attenuating ranges were fitted both in the grid and cathode circuits so that a range of inputs between 100 millivolts and 100 volts could be accommodated.

The general circuit layout is shown in Fig. 2. In the central position (2) the signal passes straight from cathode follower input via a $10 / 1$ attenuator to the output stage which has a full-wave bridge rectifier meter in its anode circuit. In the final arrangement an additional RC-coupled stage was added before the output valve in order to give sufficient gain to work at inputs down to 100 millivolts. A conventional " 1 -milliamp" full-wave bridge rectifier is employed with a 100 -microamp movement meter. The incoming wave is adjusted by the input control $\mathrm{R}_{1}$ to give full-scale deflection of the meter, and the instrument is now "calibrated" to measure the distortion of the incoming signal.

To measure the third harmonic percentage the switch is placed in position 1 when the signal passes first through the high-pass filter which attenuates the fundamental by 40 db . At the same time, since the 10/1 attenuator is now out of circuit, the gain has been increased by that amount and the full-scale deflection of the meter corresponds to only 10 per cent of the original signal. The cathode of the output valve, $V 4$, now introduces two inductances $L_{7}$, and $\mathrm{L}_{8}$, and a condenser $\mathrm{C}_{8} . \mathrm{L}_{8}$ and $\mathrm{C}_{8}$ are tuned to $2,000 \mathrm{c} / \mathrm{s}$ and introduce very considerable negative feedback at this frequency, thereby eliminating the second harmonic from the output. At $3,000 \mathrm{c} / \mathrm{s}$ the $L_{8}-C_{8}$ combination is capacitive, and $L_{7}$ is therefore added in series to resonate with it and give virtually
zero impedance : thus the third harmonic is amplified with low loss. In practice the amplifier gain is fractionally increased in the switching to make up the loss, which is about 2 db . All higher harmonics i.e., fourth, fifth, etc., suffer considerable attenuation due to negative feedback, and the output is the percentage of third harmonic distortion, the full-scale reading being 10 per cent.

For the measurement of percentage total harmonic distortion the switch is placed in position 3, when the $1,000-\mathrm{c} / \mathrm{s}$ fundamental is removed as before, but all harmonics are amplified equally and summated into the rectifier meter.

Mechanical Construction and Layout. The simplicity of the switching arrangements permits an easy and symmetrical layout of the front panel, as may be seen in Fig. 3. On the left is the oscillator unit with its output control and multiplier, and on the right the attenuator and input control to the distortion measuring unit. In the centre is a spring loaded switch, normally resting in the second position, for selection of the distortion measurement.

A normal folded chassis is employed, and, to avoid the necessity for special screening, the toroidal coils are placed underneath the chassis and clamped by the various tag panels; this form of coil is used throughout the equipment. The oscillator second harmonic rejector alone was screened.

Two double triodes ECC81 (12AT7) are used to provide the two cathode followers and the two RC stages; an EF86 is employed as oscillator with a 150B2 to stabilize its h.t. supply. No further h.t. stabilization was found necessary, but since the h.t. supply is derived from a full-wave bridge metal rectifier, a relay was added with its coil in series with the two cathode follower anodes. This normally shorts out the sensitive meter and prevents the series of bumps which occur as the valves warm up: it is also a useful transit protection.

Performance. Two of these instruments have been in continuous use for six months for setting the programme level meters of tape recorders. An endless tape is employed and recording is made at a standard input level and adjusted to correspond to maximum indicated recording level. It is immediately played back through the distortion meter. The programme level meter is successively adjusted till the required value of distortion is obtained, and it is found that with semi-skilled operators the whole of the adjustment and measurement may be completed in some two to three minutes.

For measuring distortion on audio amplifiers readings may be obtained immediately and quickly over a wide range of inputs; it is only necessary to adjust the input attenuators to full-scale deflection and then turn the switch for a direct reading. In conjunction with an output power meter a curve of distortion against output watts may be readily obtained.

The overall accuracy obtained on third harmonic measurement is $\pm 0.3$ per cent distortion and on total harmonic distortion measurement $\pm 0.5$ per cent distortion. These figures relate to the full scale of 10 per cent and are better on the lower half of the scale. Allowance can be made for the oscillator distortion if required, but since this is normally less than ${ }_{1}$ per cent for all measurements over half scale it may be neglected.

The author wishes to thank Mr. O. E. Dzierzynsiki for his assistance in carrying out the experimental work involved in developing this equipment.

# Clectronic Analogiue Computing 

Surrey of Modern Techniques

SO much has been written recently aoout the truly wonderful achievements in the field of digital computing that there has been a tendency to forget about analogue computers and to overlook the progress they have been making. Indeed there are those who would say that the analoguc computer is cutdated and dispute the need to improve it further. It is hoped in this article to show that the need still exists, and further that the digital computcr as it stands, in spite of its undoubted superiority in many cases, still has a long way to go before it can surpass the analogue device in all applications.

A useful distinction between the two classes of computer may be drawn by referring to analogue instriments on the one hand and digital muchines on the other. The instrument handles continuously varying quantities, but to an accuracy limited by the attainable perfection of its manufacture : this is not easily made hetter than about one part in 10,000 . The machine handles numbers expressed in finite digital form and is thus incapable of handling continuously variable quantities, but within this restriction is theoretically capable of any required finite degree of accuracy; in practice, accuracies of one part in a million are being achieved with machincs of the size now available.

From what has been said, it will he clear generally that if problems of an essentially arithmetic nature are posed, and exact answers required, the digital machine is called for, but that if continuously variable quantities are to be dealt with the analogue device is more suitable. The precise nature of the difference between the two classes of computer will be evident if we consider the way in which the digital machine handles its input data. First of all, it samples its data at particular instants of time. Secondly, at each sample, it recognizes only a finite number of magnitude levels, i.e., it quantizes each sample of the data. In digital computer programming, it is good practice to try to equalize the errors due to sampling and quantizing. In the analogue instrument, the counterparts of quantizing and sampling errors are respectively the imperfections in the computing components, or the static errors, and the finite response times of the computing servos, or the dynamic errors. Again, the best design requires these errors to be of equal magnitude.

Optimum balancing of the computing errors is not the only factor to be taken into account, however, in the choice of a computing system. It is sometimes essential for a natural time scale to be used in the computation, for example in certain classes of simulator.

In a simulator the system to be studied is represented by the circuits of the instrument, and the
behaviour of these circuits is studied under representative working conditions. A natural time scale is essential when actual components are to be inserted, as is done sometimes to allow a check on their performance during design. These component tests may be carried out not unly under representative circuit performance conditions but also under representative opcrating conditions of temperature, humidity, vibration and the like. It is also possible to reproduce the conditions under which a human operator is included in the servo loop.

In problems of any degree of complexity, where operation in natural time scale is desired, digital computers are sometimes too slow. Since limitation in speed of response is a characteristic of digital computers that is not always apprec:ated, it may be of interest to illustrate it by means of an example. In some work on the hehaviour of a controlled missile, a problem was posed which involved the time dependence of the missile's position and artitude in space, as well as the behaviour of certain important parts of the control system. A simplified mathematical description of the problem involved over twenty first-order simultancous differential equations, and these were solved on the Manchester University digital computer using a step-by-sten method of integration in which the step interval was adjusted according to the shortest time constant of interest in cach part of the solution.

It was found that, in a range of interest where a time interval of an cighth of a sccond was adequate, investigation of each ten seconds of the solution took just over an hour of computing time, i.e, the ratio of machine time to natural time was about 100 . In another range, where step intervals of 132 sccond were used, the ratio worked out to 1,800 . Time ratios of this order of magnitude appear to be typical for problems of this degree of complexity. In many cases, as in the example given, such a change of time scale is unimportant because the time is available, but in other cases, as we have seen in connection with the simulator, operation in the natural time scale is often essential. In some applications, where the analoguc simulator or computer is used purcly for design purposes and high accuracy is not the primary requirement, it is convenient to arrange for the computer time scale to be much less than natural and to display the response of the system to a repetitive square wave on the screen of a cathode ray tube. The results of any changes in system parameters may then be observed directly with consequent saving of a great deal of the designer's time.

- Ferranti. Led

In other problems where the data provided are continuous and high accuracy answers are required continuously, either in the natural time scale or faster than this, neither the digital nor the analogue computer may be capable of dealing with the situation adequately. It is evident that there is a need for faster digital machines, of the order of 100 to 1,000 times faster than the present Manchester machine, and also a need for electronic analogue devices of 100 to 1,000 times the accuracy at present attainable. It may be that the solution will lie in the development of a machine using a combination of both computing techniques.

## D.C. Amplifier Networks

Most of the accurate analogue computing being carried out in this country at the present time is done on electro-mechanical analogue computing instruments which use techniques developed during the war for anti-aircraft predictors and component production facilities set up at that time. While computers of this type have been considerably improved in recent years, a new and interesting field is now being opened up by the use of all-electronic devices, which require no moving parts.

In these electronic techniques the basic computing procedures of scale changing, adding, subtracting, integrating and differentiating may be carried out by d.c. amplifier networks, as shown in Fig. 1. Though balanced amplifiers have some advantages, mainly in reduced h.t. smoothing demands, a form of d.c.
amplifier often used is the unbalanced one shown in Fig. 2. It is a three-stage amplifier with resistive interstage coupling ; the odd number of stages ensures an overall sign reversal and permits feedback, while the direct coupling between stages permits operation at frequencies down to zero. As the overall gain of such a three-stage amplifier may be several tens of thousands, a small grid-to-cathode input potential suffices to produce the full output. In this case, the amplifier gain is about 50,000 and the output varies over a range of $\pm 150$ volts, so that the required grid-to-cathode voltage variation is only $\pm 3$ millivolts. Thus, in the normal operating state the input grid may be considered as being at earth potential.

A circuit incorporating a d.c. amplifier of this type which can be used either for scale changing or for sign reversing is shown in Fig. 1(a). The input voltage, $\mathrm{E}_{1}$, is applied to the input grid through a resistor, $\mathbf{R}_{1}$. Neglecting the feedback resistor, $\mathbf{R}_{0}$, for the time being, it will be seen that since the input grid is virtually at earth potential the input current, $\mathrm{I}_{1}$, is given by Ohm's law as:

$$
I_{1}=\text { input current }=\frac{\text { input voltage }}{\text { input resistance }}=\frac{E_{1}}{R_{1}}
$$

This flow of current drives the input grid (Fig. 2) positive in potential and the amplifier responds by driving the output anode voltage down. Considering now the action of the feedback resistor, $\mathbf{R}_{0}$, it is evident that it draws from the input a current of magnitude $I_{0}=E_{0} / R_{0}$ and that this current increases until it


Fig. I. D.C. amplifiers used for (a) sign reversing, (b) summation, (c) integration and (d) differentiation.


Fig. 2. Unbalanced d.c. amplifier suitable for circuits such as thase in Fig. I. equals the current $I_{1}$, when the amplifier settles down to steady state operation. Equating the expressions for the currents $I_{1}$ and $I_{0}$ gives :
$\left(E_{1} / R_{1}\right)+\left(E_{0} / R_{0}\right)=0$ or $\mathrm{E}_{0}=-\mathrm{E}_{1}\left(\mathrm{R}_{0}, \mathrm{R}_{1}\right)$, while if the input and feedback resistors are equal in value $\mathrm{E}_{0}=-\mathrm{E}_{1}$, i.e., the output is equal in magnitude to the input but has opposite polarity.

Since a large current may be taken from the output without causing a drop in the output voltage, these amplifiers are often used as buffer stages between high out-put-impedance circuits and their loads.

If two input resistors are provided, as shown in Fig. 1(b), then the current drawn off from the grid through the feedback resistor equals the sum of the input currents, i.e., the output voltage is equal to minus the sum of the input voltages. The accuracy of the summation depends, of course, on the accuracy with which the resistors can be matched;


Left: fig. 3. Four d.c. amplifiers arranged as a differential analyser.

Below: Fig. 4. Electronic function generotor in which the required function of the input voltoge is represented by the profile of a mask.
high stability wire-wound resistors are normally used. Subtraction is achieved by the addition of a voltage of opposite polarity.

Incorporating a capacitor in the feedback loop in place of the resistor, as shown in Fig. 1(c), converts the arrangement into an integrator. Since the current passed by the capacitor is proportional to the rate of change of the voltage across it, we have

$$
\mathrm{E}_{1} / \mathrm{R}_{1}+\mathrm{C} d \mathrm{E}_{0} / d \mathrm{t}=0 \text { or } \mathrm{E}_{\mathrm{n}}=-\left(1 / \mathrm{CR}_{1}\right) \mathrm{J} \mathrm{E}_{1} d \mathrm{t}
$$

Polythene capacitors are usually employed because of their high leakage resistance.

With the capacitor in the input, as shown in Fig. 1(d), the arrangement becomes a differentiating circuit, and $\mathrm{E}_{0}=-\mathrm{CR}_{0} d \mathrm{E}_{1} / d \mathrm{t}$. One drawback to differentiating amplifiers is their tendency to accentuate any irregularities in the input voltage. When a computer is being set up for the solution of a differential equation, however, it is always possible to rearrange this in the form of an integral equation. The solution then allows the use of integrators in place of differentiators.

## Differential Analyser

An arrangement of four d.c. amplifiers for solving a second-order differential equation is shown in Fig. 3. The equation solved is:

$$
\mathrm{A} d^{2} \mathrm{y} / d \mathrm{t}^{2}+\mathrm{B} d \mathrm{y} / d \mathrm{t}+\mathrm{Cy}=\mathrm{f}(\mathrm{t})
$$

and its solution is a time function, $y(t)$, representing the response of a system with two energy storage components, A and C , and an energy dissipating component, $B$, to a forcing function of time, $f(t)$. When the system is linear, $A, B$ and $C$ are constant and this is assumed in this case.

The first amplifier is connected for summing and the quantity $d^{\prime} y^{\prime}{ }^{\prime} d t^{2}$ is assumed to exist at its output. This is then fed to the second amplifier, connected as an integrator, in which $-d y / d t$ is evaluated. The third amplifier is also an integrator and gives $y$. In the first amplifier, $d^{2} y / d t^{2}$ is evaluated by summing the three terms $\left(B^{\prime} A\right) d y / d t$, ( $\left.C^{\prime} A\right) y$ and $-f(t)$, the various proportions in which they are summed being obtained by a suitable choice of the input resistors. The fourth amplifier is needed to give a sign reversal to the quantity $-d y / d \mathrm{t}$. Initital conditions, representing the levels in the two energy stores, are set into the two integrators before the forcing function is applied. This forcing function is sometimes just a step function; at other times, however, a repetitive solution is required and then a continuous square wave is applied. This enables the solution for $y$ to be presented as a steady trace on a cathode ray tube provided the frequency of the square wave is high enough to avoid flicker.
Usually it is convenient for the repetition frequency

to be that of the mains. The time scale will then be chosen to allow the whole output transient to occur in less than $1 / 50$ th second, and also to allow enough time for the flybacz and reset of initial conditions.

If a plot of the response is required, inaccuracies are introduced in trying to obtain this from the cathode ray tube screen directly. it is better to use a pen recorder, in which the accuracy is comparable with that of the computing network and in which the paper drive mechanism can have a greater speed constancy than the cathode ray tube time base. One way of doing this is by sampling the output waveform with a pulse of repetition frequency slightly different from the $50 \mathrm{c} / \mathrm{s}$, so obtaining from each successive transient a portion slightly displaced from the last. The pen recorder, being unresponsive to frequencies as high as $50 \mathrm{c} / \mathrm{s}$, will draw a smoothed version of the required response, though taking the whole of a beat period to do so.

## Function Generators

With the aid of a multiplying device, or a non-linear function generator, it is possible to solve more complicated equations in which the coefficients $\mathrm{A}, \mathrm{B}$ and C are no longer constants but functions of time. The circuits used do not differ radically from that of Fig. 3, and the method of getting the answer is very similar. The same cannot be said of the analytical solutions, as the modified equations are much more difficult to solve than the linear.

An all-electronic non-linear function generator developed recently is shown in Fig. 4. It consists of an opaque mask, having a profile representing the required function, placed on the screen of a cathode ray tube. A photocell receives light from the fluorescent spot on the screen when this is not obscured by the mask, and the output of this photocell controls the voltage applied to the vertical deflecting plates, through an amplifier. The input voltage is applied to the horizontal deflecting plates and causes the spot
to move horizontally. A bias voltage is applied through the amplifier to the vertical deflecting plates in such a sense as to cause the spot to rise until it reaches the edge of the mask. As the spot becomes unmasked, the photocell input to the amplifier opposes the bias increasingly until the amplifier output attains

(a)
a value at which the spot is prevented from rising further. The spot is thus forced to follow the mask profile, and the required function of the input voltage is generated at the output terminals of the amplifier. This device has been used in an aerial simulator to represent the functional relationship between aerial gain and offset angle, and can be used to simulate any empirical relationship of this kind.

Arrangements of biased diodes allow the representation of various non-linear saturation characteristics, such as occur in the the simulation of mechanical systems. Fig 5 (a) show's two diodes arranged to limit the output of an amplifier and the full lines in Fig. 5(b) show the resulting output characteristic. Diode $\mathrm{V}_{1}$ conducts when the output voltage, $\mathrm{E}_{0}$, exceeds the bias voltage, $E_{1}$, so $E_{1}$ corresponds to the maximum value of $E_{0}$ which can be passed. Conversely, $V_{2}$ conducts when $E_{11}$ is less than $E_{2}$, so this voltage corresponds to the minimum value of $E_{0}$. The circuit thus gives a characteristic typical of a system with position, rate or torque limits.
When a characteristic representing friction torque as a function of speed is to be generated, a circuit similar to that in Fig. 5(a) is used, but without the feedback connection to the amplifier input. The amplifier then drives its output to the limiting value for any slight departure of its input voltage from zero, in either the positive or negative direction. The resulting characteristic, which is of the desired form, is shown dotted in Fig. 5(b).

A dead zone often exists in mechanical instruments and this may be represented, in a computer, by the arrangement of Fig. 6(a). If the input signal $\theta_{1}$ is positive, it must exceed $+E$ for the diode $V_{1}$ to conduct or, if negative, be less than $-E$ for the diode $\mathrm{V}_{2}$ to conduct. A positive or negative threshold of value E is thus introduced into the output $\theta_{0}$ as shown in Fig. 6'(b).

The inclusion of a capacitor at the output, as shown in Fig. 7(a), somewhat modifies the behaviour of the circuit. The diode $V_{1}$ now conducts when $A_{1}-A_{0}>E$, and $V_{2}$ when $\theta_{1}-\theta_{0}<-E$. This results in the characteristic of Fig. 7(b), which represents the backlash in a system of gears. If a limiter is combined with this arrangement a hysteresis characteristic results.

## Approximating Characteristics

The method may be extended to the approximation of any non-lincar function to within specified limits of accuracy by a number of straight-line sections, provided the slope of the characteristic remains of the same sign throughout. If the slope of the required function varies in sign then the characteristic is obtained as the sum of a number of simpler characteristics. A high-speed electronic differential analyser at the University of Bologna has function generators of this type which permit any function to

Fig. 5. Arrangement of biased diodes to simulate a mechanical system with limits on displacement, rate or torque.

(a)

(b)

(b)

Fig. 6. Biased diodes simulating a dead zone in a mechanical system, as shown by $-E$ to $E$ in (b).

(d)

(b)

Fig. 7. Including a capocitor in the Fig. 6 circuit modifies the choracteristic to represent backlosh in a system of gears.
be set ur as a combination of ten connected line sections whose slope and length can be adjusted independently.

An application of the method is shown in Fig. 8. At low input voltages, when the diodes are all biased off, the circuit acts as a simple potential divider, but as the input voltage increases the diodes become conducting one after the other and make the effective potential divider ratio progressively less. With the circuit values and bias voltages shown, the arrangement gives a logarithmic law over a wide range, and this particular circuit has been used as an attenuator for compressing voltages from one to 1,000 volts, so allowing the voltage values in both low-and high-level regions of the signal to be read on a meter to the same percentage accuracy.

A good deal of attention has been given recently to the development of fast multipliers. Some work has been done on improving the speeds of response
of electromechanical devices, but purely electronic devices are more common.

One form of electronic multiplier uses a cathode ray tube and photocells, as shown in Fig. 9. There are the usual vertical and horizontal deffection plates in the neck of the tube and a coil is fitted round the tube in the vicinity of the horizontal defection plates. One of the input voltages, $V_{1}$, is applied to this coil and produces a magnetic field along the axis of the tube of magnitude proportional to $\mathrm{V}_{1}$. So long as the electron beam is undeflected, the electron beam current acts in the same direction as this magnetic field and does not interact with it. When, however, the second voltage, $V_{2}$, is applied to the vertical deflection plates it imparts a vertical velocity to the electron beam so that the beam current now has a component in the vertical direction of magnitude proportional to $\mathrm{V}_{2}$. This component of the beam current interacts with the axial magnetic field and produces a force on the electron beam, in the horizontal sense, of magnitude proportional to the product $V_{1} V_{2}$.

The tube face is divided by a vertically placed knife edge having a photocell on either side. This enables any horizontal deflection of the beam to be detected as a difference signal, which is amplified and applied to the horizontal deflection plates in a restoring sense. The resultant behaviour of the system is to produce a voltage across the horizontal deflection plates proportional to the product of the two input voltages.

In another clectronic multiplier, the mark space ratio of a repetitive square wave is made to depend on one voltage, and its peak-to-peak amplitude on another. The detected and smoothed output then varies as the product of the inputs. A multiplier
width and adequate suppression of side lobes. Since high accuracy is not demanded, an electronic computer may be used, and the computed radiation patterns presented on the screen of a cathode ray tube. This allows the designer to see at once the effect of changing any design parameter and speeds up what is essentially a trial and error process.

## Synthetic Radiation Pattern

The voltage which produces the synthetic radiation pattern on the c.r.t. screen is an analogue of what would be measured by a field-strength measuring set moving in a circle round the real aerial array. In the real array the waves emitted from the elements would combine to produce various maxima and minima which would be detected by the revolving measuring set. The same effect is obtained in the computer by combining a number of $450-\mathrm{kc}_{l} \mathrm{~s}$ r.f. carriers, corresponding to the emissions from the elements, after modulating their respective amplitudes and phases in different ways corresponding both to the amplitudes and phases of the currents fed to them, and to the transmission lags between them and the rotating measuring set. Thus at various points on the modulating cycle the resultant carrier output goes through maxima and minima analogous to the maxima and minima detected by the measuring set at corresponding points on its circle of rotation. Fig. 10(a) shows the apparatus for simulating the contribution of a single element, while (b) shows the combining and display system.

In order to obtain a continuous presentation of the maxima and minima on the c.r.t. screen the modulating


Fig. 8. Circuit used for compressing an input voltage range to a logarithmic scale.
may also be made from function generators arranged to give square law responses. If two such generators are arranged to evaluate $(A ; B)^{2}$ and $(A-B)^{2}$, their difference, when scaled down by a factor of four, is equal to the product AB .

Large computing assemblies, beíng expensive, are rarely justified when suited only to the solution of a special problem. The exceptions to this rule are when the problem has to be solved quickly, as in an anti-aircraft predictor, or frequently, in which case substantial savings in computing effort may rapidly outweigh the initial cost. An example in this latter class is the aerial radiation pattern computer constructed recently at the Telephone Manufacturing Company

Aerial array design involves computing a great variety of radiation patterns until one of suitable form is found, which has the right compromise between the conflicting requirements of narrow main lobe


Fig. 9. Electronic multiplier using a cathode ray tube and two photo-cells.


Fig. 10. Aerial radiation pattern computer with (a) apparatus for simulating contribution of a single element, and (b) combining and display system.
cycle is chosen to be $60 \mathrm{c} / \mathrm{s}$ and the time base is synchronized with this frequency. (This being equivalent to a rotation of the hypothetical measuring set round the array sixty times in every second.) Thus, with the azimuth angle represented along the time base and the combined carriers applied to the Y plates, the radiation pattern is presented in rectangular co-ordinates. Manual adjustment of the amplitudes and phases of the $450-\mathrm{kc} / \mathrm{s}$ carriers gives the effect of
similar adjustments to the currents in the elements of the real array, while alteration of the amplitudes and phases of the $60-\mathrm{c} / \mathrm{s}$ modulating signals gives the effect of changing the positions of these elements.

Very often the problem to be solved does not justify the building of such a large special-purpose computer. In this event it is sometimes possible to arrange for the problem to be solved on a general-purpose computer, which can be used in the solution of a large variety of problems and so be more economical. General-purpose electronic analogue computers have been built at the National Physical Laboratory, Teddington, and also at the Royal Aircraft Establishment, Farnborough.

The N.P.L. computer is intended mainly for use in the design of servo systems and comprises exponential time delays, pure time delay or distance velocity lag, balanced d.c. amplifiers, mixers, integrators, squarewave and pulse generator working at $1,600 \mathrm{c} / \mathrm{s}$, phase advancers, resonant circuits, and an on-off controller. Combinations of these clements allow most types of servo to be studied.

By contrast, the R.A.E. computer is arranged as a differential analyser, similar to those discussed previously but more complex. It comprises 16 integrators, 15 summing amplifiers, 12 curve followers or function generators, 15 multipliers, and 30 units for scaling and inserting initial conditions. The multipliers are of the variable mark'space ratio type previously described; all the other units are based on unbalanced d.c. amplifiers.

One very large use of differential analysers of this type is in the solution of flutter and vibration problems in airframe design. Since the data on which designers work are very rarely accurate to better than a few per cent, highly accurate computers are out of place, and the purely electronic type with repetitive solutions presented on the screen of a cathode tube is ideal.

## IB(1)KN IRECEIVEID

Radio Receiver Design, by K. R. Sturley, Ph.D., B.Sc., M.I.E.E. Revised second edition of Part 1 dealing with radio-frequency amplification and detection. Pp. $667+\mathrm{xx}$; Figs. 260. Price 56s. Chapman and Hall, 37, Essex Street, London, W.C.2.

Low-frequency Amplification, by Dr. N. A. J. Voorhoeve. Philips Technical Library treatise on the principles of a.f. amplifier design, with chapters on auxiliary apparatus and acoustics as applied in sound reinforcing systems. Pp. $495+\mathrm{xv}$; Figs. 479. Price 50 s . Cleaver Hume Press, Wright's Lane, Kensington, London, W.8.

Ultra High-frequency Radio Engineering, by S. A. Knight. Elements of the theory of transrnission lines, wave guides, magnetron and klystron valves, etc., with chapters on propagation and aerial systems. Pp. 256+ viii; Figs. 202. Price 21s. Sir Isaac Pitman and Sons, Parker Street, London, W.C.2.

Microwave Spectroscopy, by M. W. P. Strandberg. Theory of molecular rotational states and tneir measurement by interaction with microwave fields. Pp. 140; Figs. 15. Price 9s 6d. Methuen and Company, 36, Essex Street, London, W.C.2.
World Radio Handbook for Listeners (1954). Conıpendium of information on broadcasting and television stations, their wavelengths, interval signals, times of transmission, etc. Pp. 136 with numerous illustrations. Price 8 s 6d. Edited and published by 0 . Lund Johansen, English edition distributed by W. Dawson and Sons, Cannon House, Macklin Street, London, W.C.2.

Electrical Engineering Progress Series, Editor M. G. Say, Ph.D., M.Sc., M.I.E.E. Collections of articles on recent developments by specialist contributors.

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The above three books are published by George Newnes, Tower House, Southampton Street, London, W.C. 2 .

Mechanism of Economic Systems, by Arnold Tustin, M.Sc., M.I.E.E. Application of control system engineering methods, feedback and stabilization, to the problems of economic fluctuations. Pp. 161; Figs. 57. Price 25s. W. Heinemann, 99, Great Russell Street, London, W.C.1.

Technique de la Télévision, by A. V. J. Martin, A.M.Brit.I.R.E. Outline of television receiving technique with special reference to Continental standards. Pp. 295; Figs. 358. Société des Editions Radio, 9 Rue Jacob, Paris 6.

Gears for Small Mechanisms, by W. O. Davis. Theory and practice of designing small gears for instruments, clocks and automatic control mechanisms. Pp. 157+ix; Figs. 76. Price 25s. N.A.G. Press, 226, Latymer Court, Hammersmith, London, W.6.

# Surface-Barrier Transistors 

New Technique for Making High-frequency $\mathfrak{F u n c t i o n ~ T y p e s ~}$

1RTICLES in this journal have already described the point transistor and the junction transistor: variants, like the analogue transistor which is now reported to be approaching the practical stage, have not passed unnoticed. It is a fcature of both these basic types of transistor that they are all of the $a-b-a$ type, the junction transistors either $n-p-n$ or $p-n-p$, and the point transistors actually having small areas of $p$-germanium under the points, so that internally they are $p-n-p$ systems.

The technological problems of constructing the germanium sandwich are, like the problem of getting the jam into the doughnut, extremely difficult. The central layer must, like the jam, be extremely thin and methods used so far have not been very successful in achieving the required uniformity except for relatively thick base layers, with the corresponding limitation to low frequency operation.

A new type of junction transistor has now been announced by Philco in America, in which many of the difficulties are overcome. In some ways this new transistor reminds us of the coaxial point transistor described by Kock and Wallace'. The coaxial transistor consisted of a disc of germanium one-eighth of an inch in diameter and 0.02 in thick, in the faces of which two concave depressions were ground. The point contacts were applied coaxially, one fitting in each hollow. It seemed quite a good idea, but no more has been heard of it. The surfacebarrier transistor is also made from a small plate of germanium, cut from a single crystal of $n$-germanium and initially $0.05 \times 0.10$ in in area and 0.006 in thick. Since the cutting operation damages the crystal structure, the small blanks are etched to a thickness of 0.003 in , and a nickel contact tab is then soldered to one end.

This slab of germanium is still far too thick, and the working section must

[^4]be reduced in thickness. Here we meet Part I of the new invention. Jet electrolytic etching is used to dissolve away some of the germanium. Fig. 1 shows how two jets are played on opposite sides of the germanium blank from glass nozzles about 0.005 in in diameter. The jets are a solution of an indium salt, and a current of about 1.5 mA is passed along the jets and through the germanium, the circuit being closed through the nickel tab. Light must fall on the jet-germanium junction during the etching process, because the current is flowing through the back resistance of the jet-germanium diode, and it turns out that the effect of these combined conditions is to make the window flatten out as it gets thinner. This helps to give good high-frequency performance.

A hole is first drilled right through, to find out how long it takes on the particular blank being processed. Then the blank is moved, and the etching process carricd out, stopping when calculations show it must have left a window about 0.0002 in thick. This takes rather under two minutes.

The current through the jet-germanium diode is now reversed, so that the system becomes an electroplating unit. Indium is deposited on the surface of the germanium, and the layer is built up to give an electrode thickness of 0.0005 in . Wires are fastened to these two electrodes and the unit is mounted and sealed. The etched wafer is shown in Fig. 2, and the method of mounting in Fig. 3.

There is no forming action, no heating to produce


Above: Fig. I. Etching of germanium wafer by jets of electrolyte.

Left : Fig. 2. Fhotomicrograph of section of etched wafer. The white areas show the relative sizes of the indium electrodes.

Below: Fig. 3. Experimental mounting for surface transistor.

$I_{e}$ (milliamperes)

$V_{\mathrm{C}}=-0.5 \mathrm{~V}$ and $\mathrm{I}_{0}=-2 \mu \mathrm{~A}$ is quite satisiactory. Amplifiers giving a power gain of 18 db at $30 \mathrm{Mc} / \mathrm{s}$ and using only a 3 -volt supply have been made. The values of $x$ cut-off published for 10 units range from 36 to $49 \mathrm{Mc} / \mathrm{s}$, and the values of $x$ from 0.905 to 0.962 .
The circuit shown in Fig. 5 shows a video amplifier having a gain of 28 db and a bandwidth of $9 \mathrm{Mc} / \mathrm{s}$. It will be noticed that the collector supply is only -3 volts. The alternative junction tetrodes, which also have good high-frequency characteristics, require much high supply voltages, usually over 15 volts.
The surface-barrier transistor appears to be a stage nearer the solution of the problem of reproducible high-frequency transistors.
Acknowledgments. Figs. 1, 2 and 3 are based on Figs. 1, 3 and 4 of "Electrochemical Techniques for Fabrication of Surface-Barrier Transistors," by J. W Tiley and R. A. Williams, Proc. I.R.E., Dec., 1953, p. 1706; Fig. 4 on Fig. 4 of "Principles of the SurfaceBarrier Transistor," by W'. E. Bradley, Proc. I.R.E., Dec., 1953, p. 1702; and Fig. 5 on Fig. 2 of "Circuit Applications of Surface-Barrier Transistors," by J. B. Angell and A. P. Keiper, Proc. I.R.E., Dec., 1953, p. 1709.

Fig. 4. Earthed-emitter Characteristic of the surface-barrier transistor.

Fig. 5. Circuit of video amplifier using surface-barrier transistors.
diffusion. Here we see the fundamental difference between the transistor which results from this method of manufacture and the transistors we have encountered before. The junction with which we are concerned is that between the germanium and the indium, at the actual surface of the germanium. This is, of course, much more clearly defined than the $n-p$ junction in either a grown type or a diffused type of junction unit.

A typical set of characteristics is shown in Fig. 4. These are measured using the base current as parameter. It can easily be seen that for very small signal working the point $\mathrm{I}_{\mathrm{C}}=-0.06 \mathrm{~mA}$,


## CAMMEIRCIAL MATEIRATIIRE

Tubular Ceramic Capacitors for handling high pulse voltages (up to 5 kV peais) in radar and television. Technical bulletin No. 43, Series 2, from The Telegraph Condenser Company (Radio Division), North Acton, London, W. 3
Radio Books to be published in the first half of 1954 included in a spring list from Sir Isaac Pitman \& Sons, Parker Street, London, w.C.?.
Eddystone short-wave components and accessories for transmitting and receiving described in a 28 -page catalogue. with illustrations and rechnical data, from Stratton and Co., Alvechurch Road, W'est Heath, Birmingham, 31.
Connectors for coaxial and balanced twin cables for general purpose use, including television. Also, details of special microwave, v.h.f. and miniature types. Leaflets from 'ransradio, 138A, Cromwell Road, London, S.W.7.
Portable Electric Gramophones with Collaro pickups and motors (or record-changers). Range of four types described in a leaflet from Electric Audio Reproducers, 17, Little St Leonards, Mortlake, London, S.W.It.
Resistance Boxes, calibrating potentiometers, Wheatstonc and plug bridges, resistance standards and other measuring instruments listed on a leaflet from the Croydon Precision Instrument Company. 116, Windmill Road, Croydon, Surrey.

Tape Recorder with Truvox tape deck, crystal microphone
and 10 -in elliptical loudspeaker. Specification on a leaflet from Unitelex (London), Deptiord Bridge, London, S.E.8.
Instrument Cases of welded steel construction and chassis to fit (also in aluminium). Kange of sizes given on leaflets from Phillips \& Bonson, Pond Works, 8, Millfields Road, Hackney, London, E.5.

Television Aerial Outlet Boxes.-Instructions for wiring extensions to other rooms given on a leaflet from Aerialite, Castle Works, Stalybridge. Cheshire.
U.H.F. Turret Attenuator ( $0-3.000 \mathrm{Mc} / \mathrm{s}$ ) with six tubular pads which, in turn, are brought in line with the coaxial input and output sockets. Characteristic impedance is 50!? and attenuation steps can be from 0.1 db to 60 db . Full description on a leaflet from Stoddart Aircraft Radio Company, 66+4. Santa Monica Boulevard, Hollywood, 38, California. U.S.A.

[^5]
# Transistorized Megohmmeter 

## Compact Tzoo-range Instrument Using a Transistor E.II.T'. Generator

By P. B. HELSDON, A.M.Brit.I.R.E.

THE small size and low power consumption of a transistor e.h.t. generator makes the electrostatic type of megohmmeter an attractive proposition. The complete unit to be described measures 6 in $\times$ 4 in $\times 3$ in and operates from a small hearing-aid type battery. There are two ranges, 3 to 1,500 megohms and 35 to 22,000 megohms.

In principle, the electrostatic megohmmeter is the dual of the well known milliohmmeter or bonding meter. The basic circuit of the bonding meter is shown in Fig. 1. Its dual in Fig. 2 is the basis of the electrostatic megohmmeter. The constant voltage source equals the full-scale deflection (f.s.d.) of the meter. The voltmeter is connected across the standard resistor and is calibrated in terms of the unknown resistance. To measure very high resistances the meter must be electrostatic.

Electrostatic meters have a modified square law of deflection, so that greater sensitivity is obtained near full-scale reading. With the arrangement of Fig. 2, large values of the unknown give low voltage reading. But when the standard and unknown are interchanged as in Fig. 3, large values of the unknown give readings near full scale, so that the meter is used to best advantage. In practice the meter is switched according to the range in use. Onc is used for unknowns less than the standard and the other for those greater. The meter reads half-scalc voltage on each range when the unknown equals the standard.

In theory, the value of the standard can be as high as desired, but in practice it is limited by considerations of leakage. Meters can be obtained with leakage resistance greater than one million megohms, so that with care and a suitable standard, resistances of this order can be measured. The 750 -volt electrostatic meter used by the writer is of war surplus stock and unfortunately has a leakage of 1,750 megohms. This value has been measured on several different occasions and is assumed to be fairly stable. For the standard resistor ( $\mathrm{R}_{4 \mathrm{TD}}$ ) a value of 25 ) megohms is used.

## Source Impedance

Power requirements are modest. Allowing for the leakage, the low range requires up to $3.4 \mu \mathrm{~A}$ at 750 V and the high range up to $3.25 \mu \mathrm{~A}$ at 815 V . Regulation must be such that the effective source impedance obtained is small compared with the standard. No serious attempt has been made to analyse the regulating properties of the e.h.t. generator. Mcasurement shows the effective source impedance to be $14 \mathrm{M} \Omega$.

After adding element; for leakage and source impedance the circuit of Fig. 2 becomes as shown in


Polythene-insulated leads and wander-plug switching are employed. The meter is of the electrostatic type.


Fig. I. Basic circuit of milliohmmeter.


Fig. 2. Bosic circuit of electrostatic megohmmeter $\left(R_{\mathbf{X}}<R_{\mathrm{STD}}\right)$


Fig. 3. Circuit of fig. 2 oltered to give readings near fullscole deflection when $R_{x} ン R_{s T I U}$

Fig. 4, with an effective standar 1 resistance ( 1,750 and 250 in parallel) of $220 \mathrm{M} \Omega$.

So when $v=100, \mathrm{R}_{\mathrm{x}}=1,526 \mathrm{Ms}$ ? ; and when $v=740$, $R_{x}=3 \mathrm{M} \Omega$.

The same procedure changes Fig. 3 into Fig. 5. Applying Thévenin's theorem, the complex resistance network seen from the $R_{x}$ terminals can be reduced to a single (standard) resistance of approximately 228 ohms, in series with al open-circuit generator voltage of 750 .
From the formula in Fig. 5, when $v-100, \mathbf{R}_{\mathbf{x}}=35$ M ! ; and when $v=740, \mathrm{R}_{\mathrm{a}}=22,800 \mathrm{M}$ !.

## Oscillator Circuit

A battery driven transistor oscillator with a step-up transformer to a selenium rectifier system is used to supply the high voltage. In this oscillator circuit the transistor acts more like an astable switch, rather than as an amplifier with positive feedback. This triggering mode of oscillation is desirable since the collector dissipation is low, both during the conducting and non-conducting parts of the cycle. During the transition, however, the dissipation is high, so the switch should be as fast as possible. The Class "C" mode of operation would scem desirable for high efficiency, but every attempt at this failed because of " squegging." The peak instantaneous collector dissipation under "squegging" conditions is very high and soon damages the transistor. Satisfactory results are obtained when Class "B" operation is adopted.

The switching capabilities of a transistor can be demonstrated by the circuit of Fig. 6. The ohm-


Fig. 4. Practical equivalent circuit of Fig. 2 for low resistance range.


Above: Fig. 5. Practical equivalent circuit of Fig. 3 for high resistance range.

Left: Fig. 6. Circuit for demonstrating the function of a transistor as a bistable switch.
meter should have an internal $9-\mathrm{V}$ battery and a halfscale reading of about 5,000 ohms. If the bias resistor has the right value the collector-emitter resistance indicated with base floating can be either high or low.

If the base is momentarily connected to the emitter, the transistor triggers to high resistance (off), but if the base is momentarily connected to the collector the transistor triggers to low resistance (on.) This constitutes a bistable switch. A pair of headphones added as in Fig. 7 makes a simple astable switch.

A better oscillator circuit is shown in Fig. 8. The base bias resistance $R_{2}$ should be increased until oscillations just start and then the emitter series resistance $R_{1}$ adjusted for maximum output. The tap on the inductor should be about ${ }_{8}^{1}$ of the total down from the base.

The transistor can be regarded as a switch which connects the tuned circuit to the battery throughout each negative half cycle. The peak-to-peak voltage developed across the tuned circuit is then twice the battery voltage. The collector has to withstand this double battery voltage at the peak of the non-conducting half cycle. As 33 V seems a safe peak collector voltage, the battery must be limited to $16!2 \mathrm{~V}$.

In practice the $16,2 \mathrm{~V}$ to the transistor is obtained from a $30-\mathrm{V}$ battery (Type B105) through a decoupled series resistance. This resistance is provided by a fixed safety resistor and a variable resistor ganged to the on-off switch. The variable resistor compensates for battery ageing and gives a precise control of the e.h.t. voltage. In addition it reduces the current to a low value before switching off, so that dangerous inductive surges are avoided. The complete circuit is shown in Fig. 9.

The oscillator is sometimes reluctant to start in cold weather. A transistor, unlike a valve, has no gain at zero bias. The emitter is biased only by the small collector leakage current passing through the base bias resistor $\mathbf{R}_{2}$. When the crystal is cold this current is small, due to the negative coefficient of resistance with temperature of high-purity germanium. After switching on, the collector current slowly rises as the crystal warms up. Seconds or even minutes may pass before the circuit bursts into oscillation. Quick starting can be obtained by connecting a fixed resistor of about $39 \mathrm{k} \Omega$ from collector to base. This is wasteful of current and also loads the tuned circuit. The method adopted is to connect this resistance by way of a pushbutton starting switch. Lightly loaded oscillators can be made to start by suitably phasing the switching-on surge, but the rectifier load in the present case initially constitutes a virtual short circuit on the tuned circuit and prevents " kick " starting.

## Auto Transformer Design

To obtain 830 volts d.c. from the peak-to-peak rectifier system, the inductance must be made into an auto transformer with a step-up ratio of $83033=25 / 1$ approx. The battery voltage adds in series giving the required total output of 860 volts on no load.

The auto transformer is wound on a four-section polystyrene former with a small pair of "Ferroxcube" E cores (FX 1105/A4) un-gapped. A total of 1250 turns of $42 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. vinyl acetal enamelled grade M (medium thickness) wire is tapped at 5 turns for the emitter and 50 turns for the battery positive. Total inductance is 4.25 H with a Q of 50 at the self-resonant frequency of $32 \mathrm{kc} / \mathrm{s}$. The extra stray capacitance of


Fig. 7. High resistance headphones in the base circuit demonstrate the astable switching function of the transistor.


Fig. 8. Sine-wave oscillotor using the astable switching circuit.
the rectificr system reduces the frequency of oscillation to about $20 \mathrm{kc} / \mathrm{s}$. The waveform is a distorted sine wave on which is superimposed damped oscillations at a higher frequency. These high-frequency ripples are produced during the switching transitions and are caused by the several leakage inductances resonating with stray capacitance. It is important to keep leakage inductance to a minimum, so that the switching transients do not over-dissipate the collector.

## Reduced Current Consumption

A large percentage of the power loss is due to collector-to-base leakage within the transistor. The low current consumption of 2.2 mA was obtained by forming a transistor to an $\mathrm{I}_{1 \cdot 0}$ (collector current for zero emitter current) of 0.6 mA at 30 volts. The average commercial point-contact transistor has an $1_{1.0}$ of 1 to 2 mA at $\mathrm{E}_{1} \cdot=30$ volts.

The complete unit is contained in a wooden box with high-grade insulation where necessary. The polythene test leads are brought out to crocodile clips to avoid the use of terminals. Meter switching is by plug and socket, since few switches have sufficiently good insulation.

On the low range the zero is set by shorting the test leads together and adjusting the vari-

Fig. 9. Complete circuit diagram of two-range megohmmeter with transistor e.h.t. generator.


# WORLD OF WIRELESS 

Death of Major Armstrong Radio Fuse Awards More Band 1 Stations:

F.M. Broadeasting in U.S.

RADIO INSTALLATION on the floating whaling factory "Abraham Larsen" which is the headquarters vessel for a whaling fleet operating in the Antarctic. The equipment was supplied by Redifon.


## F.M. Pioneer's Death

WITH REGRET we record the death in tragic circumstances in New York on February Ist of Major Edwin H. Armstrong, the f.m. pioncer. Born in December, 1890, Edwin Armstrong graduated in electrical engineering from Columbia University at the age of 23 and had worked there ever since, becoming Professor of Electrical Engincering in 1934.
Professor Armstrong will be remembered not only for his work on f.m. but for his earlier work on the regenerative and super-regenerative receiving circuits and also the principle of the super-heterodyne. He contributed a letter on the significance of regeneration to our January issue. His latest work was concerned with a multiplex system of f.m. transmission, a brief description of which will be found elsewhere in this issue.

## More Radio Fuse Auards

IN ADDITION to the $£ 20,000$ awarded by the Ministry of Supply to Pye, Ltd., for the development during the war of the proximity fuse and the No. 19 tank set (see our January issue), recommendations of awards to seven claimants have been made by the Royal Commission on Awards to Inventors. Four of the claimantsH. Cobden-Turner, G. M. Tomlin and L. Rollin, all of Salford Electrical Instruments, Lid., and W. H. B. Lord, late of that company-have been awarded jointly $£ 11,500$. Individual awards have also been made to A. Stratton ( $£ 2,000$ ), N. Coles ( $£ 750$ ) and G. A. Whitficld (£750), who are at R.A.E., Farnborough.

## New TV Stations

AMENDED PLANS for the completion of the B.B.C.'s single-programme chain of television stations were recently announced by the Assistant P.M.G. in the House. Whereas it was originally planned for eight of the twenty stations to go into Band 3, six of them will now bc accommodated in Band 1, although this is contrary to the plan drawn up at Stockholm.
The two channels in Band 3 (186-191 and 191-196 $\mathrm{Mc} / \mathrm{s}$ ) to be made available for television will now be used for the proposed competitive stations. Channel 9 ( $191-196 \mathrm{Mc} / \mathrm{s}$ ) will be assigned to London.
The six B.B.C. stations referred to above will be at, or near, Norwich, Dover, Inverness, Londonderry, Towyn and Carlisle They will each use horizontal polarization. The site for the Norwich station, where it is hoped to have a temporary $0.5-\mathrm{kW}$ transmitter working within
twelve months, is at Tacolneston, some 10 miles south west of the city.
A temporary station is also to be brought into service this year at Redmoss to serve Aberdeen until the permanent transmitter is installed at Core Hill.

The much-debated North Hessary Tor site for the Devon-Cornwall television transmitter has now been approved by the Minister of Housing. This $5-\mathrm{kW}$ station will operate in Channel 2.

## Mobile Radio

A WRIT has been served on the Post Office by the Colchester engineering firm, Davey, Paxman \& Company, alleging that the fees collected for licences for the operation of their mobile radio transmitters are not lawfully chargeable. They also claim that some of the conditions regarding frequencies and power, etc., laid down in the licence, are contrary to the provisions of the Wircless Telegraphy Act.

The outcome of the action will be of interest to all civil users of mobile radio.

## Whither F.M. Broadcasting?

IN AMERICA, the home of f.m. broadcasting, there are indications that all is not well in the f.m. camp. A member of the Federal Communications Commission recently stated that "if the broadcast industry does not take some steps to increase the utilization of the f.m. frequencies" he had no hesitancy in saying that in the public interest he would have difficulty in retaining the whole of the $88-108-\mathrm{Mc} / \mathrm{s}$ band for f.m. broadcasting should the Commission be asked to make use of part of it to relieve congestion in other services.
According to the latest returns available from the F.C.C., the number of f.m. broadcasting stations has decreased from some 750 to 560 during the past four years.

## Radio Exports Up

A RECORD YEAR for radio exports was reported in the Board of Trade returns for 1953. The total value of exports of radio equipment of all kinds was $£ 25,761,818$, compared with the 1952 figure of $£ 24,495,950$.

The contributions by the four sections of the industry are given below in £M, together with (in parentheses) the 1952 figure:-
Transmitters and radio navigationa! aids ....... 11.001 (7.963) Components and sound reproducing equipment ... $\quad 9.042$ (8.481) Domestic receivers .................... ....... 3.517 (4.435) Valves and cathode-ray tubes …..................... 2.200 (3.615)

## PERSONAIITIES

H. O. Sampson, who has been appointed head of technical operations, B.B.C. Television Siudios, joined the Alexandra Palace staff in 1936 and remained there until the television service was suspended at the outbreak of the war. During the war he served at several of the Corporation's high-power transmitting stations and with the London Recording Unis. He returned to the television service in 1946.

Eric E. Jones, who was for some vears in charge of the relecommunications Division of the Philips group of companies in this country and has, since 1949, been commercial manager of Savage and Parsons, Ltd., recording equipment manufacturers of Watford, has taken over the direction of the commercial activities of the Solartron Electronic group of companies.
In our note in the last issue recording the appointment of Rear-Admiral (L) C. P. Clarke, C.B., D.S.O., R.N. (ret.), as a Knight Commander of the British Empire (K.B.E.) in the New Year Honours, we omitted to mention his position at the Admiralty and his radio associations. At present director of the Naval Electrical Department, Rear-Admiral Clarke has been a vice-president of the Brit.I.R.E. since October, 1952.

Hugo Gernsback, editor of our New York contemporary Radio-Electronics, has been appointed Officer of the Oaken Crown by H.R.H. The Grand Duchess Charlotte of Luxembourg. The award was made to Mr. Gernsback, who was born in Luxembourg, "for his meritorious service to science."

## (lUR AUTIIIIRS

J. R. Brinkley, who has been technical director of Pye Telecommunications, Ltd., since 1949, writes on the problems of frequency planning on page 103 of this issuc. He received his early technical training in the Telepnone Department of the Post Office, and after a ycar as inspector in the Enginecr-in-Chief's Office went to the Radio Branch at Dollis Hill in 1939. He was later seconded to the Home Office, where he was responsible for the develonment of police v.h.f. radio systems and in particular for the introduction of the multicarrier mohile radio sustem

J. R. BRINKLEY

T. D. CONWAY
T. David Conway, author of the article on the measurement of harmonic distortion, has been chief engineer of Grundig (Gt. Britain), Lid., since its formation in 1952. He foined Grundig from the Ministry of Supply, where he had been an instrument engineer for two vears. Throughout the War he was with Ultra Electric as a radar engineer and from 1945 to 1947 he was concerned with the development and testing of ceramic and mica condensers at the United Insulator Company's works. Mr. Conway was with Standard Telephones and Cables as a factory valve engineer from 1947-1950.
R. B. Quarmby, contributor of the analogue computing article in this issuc, became a probationary college apprentice at Metropolitan Vickers in 1936. A year later he went to Manchester University, where he graduated in 1940, and then spent some time on extra-nural research on the properties of polythene dielectrics, etc. for the Ministry of Supply. Wartime work at the Royal Militarv College of Science was followed by a year's lectureship in electrical engineering and electronics at the University of Capetown and in 1950 he joined the staff of Ferranti, Ltd., where he has been working on the design and construction of analogue computers and simulators for use in the design of guided weapons.
D. H. C. Scholes, who contributes an article on frequencvshift radio telegraphy in this issue, has been with the Plessey Company since 1946, where he is now chief engineer and salcs manager of the Telecommunications Division. Fof a short while at the beginning of the war he worked at the Royal Aircraft Establishment, Farnborough; then, in 1941, joined the Royal Navy as an air enginecring officer and was engaged in various theatres of war on enginecring duties concerned with airborne radio and radar equipment. Lt. Cdr. Scholes was at the Admiralty for a year after the war. For seven years before his war service he was with the Marconi Company.

## I. B BRIEF

Broadcast Receiving Licences current in the United Kingdom at the end of December totalled $13,268,270$, including $2,956,8+6$ for television and 206,348 for car radio sets. The month's increase in television licences was $110,619$.

French Components Show.-The annual Paris exhibition of radio components, valves, accessories and measuring equipment will be held in the Parc des Expositions, Port de Versailles, from March 12 th to 16 th .

A course on Point-to-Point Radio Services, which is open to overseas senior planning and operating engineers and telecommunications administrators, is being conducted in London from June 27th to July l0th by the British Council in conjunction with the Engineering Department of the Post Office. Information on the course, for which there is a limited number of vacancies, can be obtained from overseas offices of the British Cruncil, or from 65, Davies Sirect, London, W'.1. 'lhe fee is $£ 38$.

Quick Work.-At the time of the Comet airliner crash near Elba, lye engincers were developing a new underwater television camera for operation at a depth of 500 fathoms. Although the equipment had hardly gone beyond the drawingboard stage, the camera, in a case for operation at the estimated depth of the Comet (some 40 fathoms), was ready within six days. It was then learned that the wreck was probably 100 fathoms below the surface. A new case was designed and completed within seven davs and flown to Italy for use by the Admiralty in the search for the wreck.

Bronze Medallist. - The City and Guilds' Bronze Medal for the best student in Great Britain in the 1953 Intermediate Radio Servicing Examination has been awarded to John .McCubbin, a Glaswegian, who is employed by James Anderson \& Son (Glasgow), Lid. He is a fourth-year student at Allan G'en's Further Education Centre, Glasgow, studying for the R.'.E.B. final certificate.
F.M. for Marine V.H.F.-On the question of the proposed change to f.m. for international maritime mobile radio services, the Radio Communication and Electronic Engineering Association states in its annual report that it has advised the Post Office that the maiority of its members favour the change provided there are suitable safeguards 10 minimize dislocation of services, especially during the changeover.
R.C.E.E.A. Council.-The following member-firms of the Radio Communication and Electronic Engineering Association were elected to the Council of the organization for 1954 (the names of the companies' representatives are given in parentheses): B.T.-H. (V. M. Roberts). Noma Ralor (C. H. T. Johnson), E.M.I. (S.J. Preston), G.E.C. (M. M. Macqueen), Kelvin \& Hughes (C. G. W'hite), Marconi's (F. S. Mockford), Metrovick (L. H. J. Phillips), Mullard (Г. E. Goldup), Murphy (K. S. Davies), Plessey (P. D. Canning), Redifon (B. St. J. Sadler) and S.T.C. (L. T. Hinton). C. G. White, general sales manager and director of Kelvin Hughes (Marine), I.td., who has been vice-chairman of the Association for the past two vears, has been elected chairman in succession to T. E. Goldup.

## INDUSTRIAI NEWS

Moroccan Television.-The first television station in North Africa, established by a private company at Casablanca, has been cyuipped with Pye, a camera chain, telecine camera and vision and sound transmitters. The station operates on the French 819-line standards.

Hi-Fi Agencies.-Enquiries from Hollywood and Hong Kong in act as representatives of U.K. manufacturers of high-fidelity sound recording and reproducing equipment have been received through the Export Services Branch of the Board of Trade. Interested firms should write direct to Gordon .Igencies, 1506, North-Western Avenue, Hollywond 27, U.S.A.,
and Scientific Service Company, 447, Alexandra House, Hong Kong. The Board of Trade, Lacon House, Theobalds Road, London, W.C.1, should also be notified, quoting references ESB/1614/54 (Hollywood) and ESB/1897/54 (Hong Kong). An enquiry has also been received from Long Island, New York, for the agency for a tape recorder. The firm concerned is Rek-O-Kut Company, Inc., 38-01, Queens Boulevard, Long Island Citv 1, New York. The B.o.T. reference is ESB/1103/54.

A contract for the supply of a yuantity of airfield control radar equipment has been awaided to Decea Radar by the Ministry of Supply. 'The equipment, which is the air-transportable version of the Decca 424 described in our November issue, will be used by the R.A.F.
"The Ship of the Year," as the new P. \& O. liner Arcadia was described at her launching last May, is being fitted with radio communication and navigational equipment by the Marconi Marine Company. In addition, a comprehensive sound-reproducing system, including 156 loudspeakers fed by seven power amplifiers, each having an output of 60 wairs, has been installed. Four sources of programmesmicrophones, broadcast receiver, wire recorder and record player-are available and switching permits selected groups of loudspeakers to be fed simultancously from the four separate sources.
Sound amplitying equipment provided by Hadley Sound Equipments, Lid., of Smethwick. Birmingham, has been installed in the Roman Catholic cathedral in Calabar, Southern Nigeria.
W. F. Kandail, B.Sc.(Eng.), M.I.E.E., director of the Telegraph Construction and Maintenance Company, is on a two months' wisit to the company's branches and representatives in the Far East and Australasia.
New premises have been opened by the G.E.C. at Magnet House, Mincing Lane, Blackburn. The depor is equipped for the supply and servicing of the company's radio and television equipment.
British Insulated Callender's Construction Company, manufacturers of radio masts and the like, recently moved to 30 , Leicester Square, London, W.C. 2 (Tel.: Trafalgar 7777). The central administrative offices of the parent company, British Insulated Callender's Cables, remain at 21. Bloomsbury Street, London, W.C. 1 ('Tel.: Museum 1600).
Goldring Manufacturing Company (Gt. Britain), Ltd., is the name of the company recently formed to conduct the business of Erwin Scharf, manufacturer of the Goldring pickup, and associated concerns The company will operate from the present address: 49-51a, de Beauvoir Rnad, Kingsland Road, London, N.I (Tel.: Clissold 3434).
Excel Sound Services, Ltd., tape-recorder and amplifier manufacturers. of Shipley, Yorks, have moved to Celsonic W'orks. Garficld Avenue, Bradford, 8, Yorks (Tel.: Bradford 45027).

## IUEETIN(SS

## Institution of Electrical Engineers

Radio Section.-"A Study of some of the Properties of Matter affecting Valve Reliability" by E. A. O'Donnell Roberts, M.Sc., Ph.D., on March 10th.
"Colour Television" by C. J. Hirsch on March 22 nd.
Both the Radio Section meetings will be held at 5.30 at Savoy Place, London, W'.C. 2.
Cambrilge Radio Group.-" Computing Machines" by T. Kilburn, M.A., Ph.D., at 8.15 on March 9th at the Cavendish Laboratory, Cambridge.
Mersey and North Wales Cintre-Faraday Lecture "ElectroHeat and Prosperity" by O. W. Humphreys, B.Sc., at 6.45 on March 25th at the Philharmonic Hall, Liverpool.
North-Eastern Radio and Measurements Group.-"A Study of some of the Properties of Matter affecting Valve Reliability ${ }^{\text {s* }}$ by E. A. O'Donnell Roberts, M.Sc., Ph.D., at 6.15 on March 15th at King's College, Newcastle-upon-Tyne.
North-W"estern Centre. -Faraday Lecture "Electro-Heat and Prosperity" by O. W. Humphreys, B.Sc., at 7.30 on March 23 rd at the Free Trade Hall, Manchester.
North-W'cstern Radio Group.-"Distributed Amplifiers" by W. S. Percival, B.Sc., at 6.30 on March 17th at the Engineers' Club, Albert Square, Manchester.

North Lancashire Sub-Centre.-" Some Aspects of the Design of V.H.F. Mobile Kadio Systems" by E. P. Fairbairn, B.Sc., at 7.15 on March 10th at the N.W. Electricity Board, North Road, Lancaster.
South-East Scotland Sub-Centre--" Some Aspects of the Design of V.H.F. Mobile Radio Systems" by E. P. Fairbairn, B.Sc., at 7.0 on March 16th at the Heriot-Watt College. Edinburgh.

South Midland Radio Group.-" Some Aspects of the Design of V.H.F. Mobile Radio Systems" by E. P. Fairbairn, B.Sc.. at 6.0 on March 22 nd at the James Watt Memorial Institute, Great Charles Street, Birmingham.

Southern Centre.-"Technical Arrangements for the Sound and Television Broadcasts of the Coronation Ceremonies" at 6.30 on March 3rd at the S.E.E.B. Headquarters, Hove, bv W. S. Proctor, M. J. L. Pulling, M.A., and F. W'illiams, B.Sc.
"Demonstrations of Synthetic Speech" by W. Lawrence, M.A., and R. A. Eades, B.Sc., at 6,30 on March 5th at the South Dorset Technical College, Weymouth.

W'estern Centre.-"Telemeturing for System Operation" by R. H. Dunn, B.Sc., and C. H. Chambers at 6.0 on March 8th at the Electricity Offices, Colston Avenue, Bristol.

South-Western Sub-Centre.-" Colour Television: Some Subjective and Objective Aspacts of Colour Rendering" by G. T. Winch at 3.0 on March 10th at the Rougement Hotel, Exeter.

Oxford District.--" Modern Trends in Television" by G. G. Gouriet, B.Sc., at 7.0 on March 17th at the Southern Electricity Board, 37, George Street, Oxford.

London Students' Section.--Visit to G.P.O. Research Station, Dollis Hill, N.W.2, at 2.15 on March 3rd.
"Metallic Resistance at High Frequency" by A. D. Stevens at 7.0 on March 3rd at the Public Library, Chelmsford.
"Servo Mechanisms" by Capt. R. A. Middleton, R.E.M.E., at 7.0 on March 30th at the Drill Hall, 185, London Road, Chelmsford.

## British Institution of Radio Engineers

London Sectiom.--"Radio Astronomy" by R. Hanbury Brown (Jodrell Bank Experimental Station) at 6.30 on March 31 st at the London School of Hygiene and Tropical Medicine, Keppel Strcet, London, W.C.I.

Scottish Section.-" 'The Accustic Design and Measurement of Buildings" by H. C. Watson (Newall's Insulation Co.) at 7.0 on March lith at the Institution of Engineers and Shipbuilders, 39 Elmbank Crescent. Glasgow, C.2, and at 7.0 on March 12th in the Department of Natural Philosophy, The University, Edinburgh.
North-W'estern Section.-"Colour Television" by G. B. Townsend (G.E.C. Research Laboratorics) at 7.0 on March 11 th at the College of Technology, Manchester.

West Midlands Section.- Industrial Applications of Electronic Instruments" by A. G. Wray, M.A.; (Marconi Instruments) at 7.15 on March 23 rd at the Wolverhampton and Staffordshire Technical College, Wulfruna Street, Wolverhampton.

## British Sound Recording Association

London.-" High Quaity Microphones: The Assessment of Performance" by D. E. L. Shorter, B.Sc., at 7.0 on March 19th at the Royal Society of Arts, John Adam Street, London, w.C.2.

Manchester Centre.-" Disc Recording with special reference to Long Playing" by A. R. Sugden at 7.30 on March 15th at the Engineers: Club, Albert Square, Manchester.

## Television Society

L.ondon.-"Trick Effects in 'Television Production" by D. R. Campbell (B.B.C.) on March 12th
"An Industrial Television Channel" by R. J. Boddy and C. D. (ardner (E.M.I.) on March 25 th.

Both the London meetings of the Television Society will be held at 7.0 at the Cinematograph Exhibitors' Association, 164, Shaftesbury Avenue, London, W.C. 2.
Leicester Centre. "Transistors and Other Crystal Valves " by D. D. Jones (G.E.C. Research Labs.) at 7.0 on March 29 th in Room 45, The College of Art and Technology, The Newarkes, Leicester.

## Institution of Electronics

North-W'estern Branch.-" Semi-conductors " by R. Cooper. Ph.D., M.Sc.. (Liverponl University), at 7.0 on March 26th in the Reynolds Hal!, College of Technology, Manchester.

## Radio Society of Great Britain

"'Trustworthy' Valves and Their Manufacture" by G. P. Thwaites, B.Sc.(Eng.), at 6.30 on March 26th at the I.E.E.. Savoy Place, London, W.C.2.

## Engineers' Guild

Metropolitan Branch.-"Television Broadcasting and the Engineer" by M. J. L. Pulling, M.A., (B.B.C.) at 6.0 on March th at Caxton Hall, Westminster, S.W.1.

## Institution of Production Engineers

Liverpool.-"Induction Hardening" by R. H. Barfield, D.Sc., at 7.30 on March 24th at the Adelphi Hotel, Lime Street. Liverpool.

## LETTERS TO TIIE EDITOR

The Editor does not necessarily endorse the opinions pxpressed by his correspondents

## Regenerative Single-Transistor Receiver

WE read with great interest the article on transistor receiving circuits by B. R. Bettridge in your January issue, having been experimenting in the same field with the GETI for some time. We cannot add to his useful information on 2 -stage circuits for local station loudspeaker reception, but some of your readers who may be able to afford only one of these treasures to begin with, and are interested in getting the greatest sensitivity with a single stage, may find Mr. Bettridge's warning against positive feedback in the base circuit unduly discouraging. With the circuit shown in the accompanying diagram, controllable bias and r.f. feedback are combined, and the danger of destructive oscillation is avoided by the limiting resistor in the emitter circuit.
In operation with a very poor "tree aerial" the circuit brings in continental stations at good signal strength, once you get used to co-ordinating the three variable controls. Moreover, with patient resolution of carriers, in which the variable base resistance is critical, many American stations can be received at stable listening volume. New York and New Orleans are particularly easy; both are in what seems to be the most favourable frequency band, around $870 \mathrm{kc} / \mathrm{s}$, though others between 600 and $1,200 \mathrm{kc} / \mathrm{s}$ come in well; talks and plays can be comfortably followed between 2 and $3 \mathrm{a} . \mathrm{m}$. The total consumption is only 18 mW .
Youngsters who have the patience to experiment with transistors in this way may thus get a taste of the thrills

which two of your carliest readers recall, sitting up at night with one of the wonder sets of the period to get the first whisper from across the Atlantic.
W. GREY WALTER, KARL WALTER.

Bristol, 9.

## Symbols

THE symbol for the current source in the technical literature is usually chosen to be an arrow as shown at (a) in the accompanying diagram. I would suggest that the symbol (b) should be adopted, this being more logical as well as exactly analogous to the symbol for a voltage source commonly used and shown at (c).

The ideal constant-current generator is assumed to have no shunt admittance, just as the ideal constant-voltage generator has no series impedance. In practical circuits internal shunt admittance and internal series impedance appear connected in their appropriate places. For example, the equivalent circuit for a transistor shown at (b) can easily be correlated with that at (c). There is the additional

advantage that the current source appears connected to the rest of the circuit which is not the case with the arrow symbol of (a).

In this connection, it is felt that the introduction of new symbols for the triode transistor is rather superfluous, bearing in mind that the symbol originally introduced by W. Shockley and shown at (d) is quite adequate.
F. OAKES.

Ferguson Radio Corporation, Enfield, Middx.

## "C..R. Tube Safory"

THE articie in your January issue considers the importance of correctly choosing the time constants associated with the cathode and grid of a television tube in order to prevent the bias going positive immediately after the receiver is switched on or off. I cannot however, altogether agree with $W$. Tusting's approach to this matter. In modern television receivers in which the e.h.t. potential for the tube is derived from an overwind on the line transformer and no e.h.t. bleeder resistor is used, the time constants in question are normally chosen so that the tube passes beam current immediately after switch-off. This ensures that the e.h.t. smoothing capacitor is discharged via the tube whilst the time bases are collapsing. If in a receiver which has a permanent magnet focussing unit the grid potential of the tube decays more quickly than that of the cathode, the tube will be biased off immediately the set is switched off. Then as both potentials approach zero the cathode may still be emitting and a bright focused spot will appear at the centre of the screen. To obviate this possibility many receiver designers prefer to arrange the grid and cathode circuit time constants so that the tube bias decreases immediately after the set is switched off.

However, if in meeting this requirement or because of some other circuit condition there is a danger of the tube grid becoming positive with respect to the cathode, the positive bias may be limited by means of a series grid resistor. The value of this resistor will depend upon the grid-cathode characteristic of the tube in the positive grid region and the extent to which the source of bias voltage travels positively. In general a resistor of $22 \mathrm{k} \Omega$ will serve to limit the positive grid excursion to +1 volt, a value which is generally accepted by tube manufacturers as being allowable for a short period immediately after switch-off. The general circuit arrangement is shown in

the accompanying diagram, the time constant ClR1 serving to hold the grid voltage up to ensure that the e.h.t. capacitor is discharged through the tube. The grid limiting resistor is R 2 .
D. A. WARD.

Mullard, Lid., London, W.C.2.

## " Weathering of Polythene"

I HAVE read with interest the short note in your December issue (page 570) which is a précis of a Ministry of Supply document " Reports on Plastics in the Tropics : 2-Polythene."
Neither your note nor the eeport itself brings out the point that the addition of 0.1 per cent of carbon black is intended only to provide identification of the cores and is not intended to provide protection against ultra-violet light. The approved practice, for this use, established over a number of years, is to include 2 per cent of carbon black of suitable grade and well dispersed in the polythene. This material would be found to possess resistance against the effect of ultra-violet light adequate for commercial use and immeasurably superior to most natural or synthetic cable materials. The addition of this quantity of carbon black does, of course, impair the initial power factor of the material at all frequencies, but it will be found that the power factor is quite stable on exposure to tropical sunshine. Observation of power factor during the course of testing is valuable as a sensitive method of revealing physical or chemical changes in the material.

Our tests indicate that the power factor of polythene with 2 per cent carbon black is not higher than $30 \times 10^{-4}$ at frequencjes up to $3000 \mathrm{Mc} / \mathrm{s}$. except in the band $3-15 \mathrm{Mc} / \mathrm{s}$. where it rises to a maximum of about $65 \times 10^{-4}$ at $5 \mathrm{Mc} / \mathrm{s}$. The commonest application where polythene is exposed to sunlight is when used as a cable sheath and here the power-factor is rarely of direct interest.

It would perhaps be unwise to deduce firm estimates of the life of this material from accelerated ageing tests, but it may be mentioned that exposure tests in India for periods which have so far attained four years have failed to reveal any change whatsoever in its electrical or physical properties.

One further characteristic of polythene is its resistance to oxidation and this matter is not considered in great detail in the Ministry of Supply report. The antioxidant most generally used with polythene is not well suited to continued operation at high temperatures on account of its volatility. However, other suitable antioxidants are available for use where the polythene is to be used as a black cable sheath exposed to tropical sun.

In view of your conclusion in the penultimate paragraph that polythene cables are unsuitable for desert conditions, I should be grateful if you would give this contrary opinion adequate publicity.

Although the work reported in the M.o.S. document is
extremely valuable and no doubt correct in its facts, it was realized in many quarters that it was open to misunderstanding, and representations have already been made to the Department concerned
R. C. MILINNER.

The Telegraph Construction and Maintenance Co., L.ondon, S.E. 10.

## "Measuring Von-lineariv"

IN my article in the February Wireless World I find that Appen lix 1 is full of mistakes which I overlooked because the final equation (5) gives a correct numerical answer. For my gross carelessness in this matter I hope readers will accept my apology. The equations in Appendix 1 should read as follows:-


The above equation (6) also appears in the main text as equation (5). Fortunately, these mistakes do not effect the body of the argument. There is also one misprint in Appendix 2-a redundant capital "I" has been inserted at the end of the last term in equation (7).

Camberley, Surrey.
1). C. PRESSEY

## Short-wave Conditions

Predictions for March

THE full-line curves given here indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during March.

Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.


- FREQUENCY EELOW WHICH COMMUNICATION SHOULO BE POSSIBLE OM ALL UNDISTUREED DAYS
-     -         - PREOICTED AVERAGE MAXIMUM USABLE FREQUENCY
........... FAEQUENCY BELOW WHICH COMMUNICATION SHOULD AE POSSIBLE FOR $25 \%$ OF THE TOTAL TIME


# Frequency Shiit Radio Telegraphy 

Modifying Radio Equipment<br>for Teleprinter Operation

By 1). II. C. NCHOLEN*

READERS have already been given a very adequate introduction to this subject in an article by Thomas Roddam, $\dagger$ and it will consequently be possible to devote our present attention to considering in rather more detail some of the problems and latest developments, particularly in connection with the receiving end of the circuit.

The various techniques for f.s.k. (frequency shift keying) reception seem to be surrounded by as much controversy as the f.m.-a.m. question, and many systems which seem theoretically sound are discredited by at least a proportion of the practical results. The whole question is complicated by the many factors which enter into long-distance radio working, and it is not possible with present experience to draw any firm conclusions as to the best system. However, some of the main advantages and shortcomings of the two principal reception systems are fairly firmly established and these will be discussed in the appropriate place.

Whatever system of resolving the teleprinter signal is employed, it is now firmly established that while for on-off c.w. operation, of either auto-morse or teleprinter, triple space-diversity reception is desirable, in the case of f.s.k. triple diversity shows no improvement over dual. It now appears that the most effective receiver is a dual diversity system where the channel giving the poorer signal is completely suppressed. The switching system should have a very rapid response and should be operated by very small level differences even during deep fades on both channels.

## Teleprinter Characteristics

In discussing radio teleprinter systems, in radio publications generally little consideration is given to the characteristics of the machine itself. The writer feels that this is unfortunate, in that the whole design of such systems is directed to the one end of producing a signal which will enable the teleprinter to operate without error. For this reason the electronics is inextricably bound up with the peculiarities of the machine itself, which must be understood before the problems connected with the radio side can be appreciated. As the teleprinter is a specialized machine it seems probable that many readers, while knowledgeable in the electronic field, may not be so well informed regarding the teleprinter, and the writer asks the indulgence of those skilled in the art during a brief digression on the operation of the machine.

Operation of any key on the sending machine causes it to produce a signal of varying polarity consisting of marks and spaces lasting for 150 milliseconds. This comprises a "start" signal followed by a $100-\mathrm{msec}$. period during which the polarity may be any combina-

[^6]tion of marks and spaces, the change-over from mark to space (if any) taking place at $20-\mathrm{msec}$. intervals. These five $20-\mathrm{msec}$. periods are the five units of the " five unit" code, and each letter or symbol on the keyboard has a corresponding combination of marks and spaces. The complete signal is terminated by a $30-$ msec. " stop" signal of opposite polarity to the " start" signal. The purpose of the start and stop signals is to control the mechanism of the receiving teleprinter. The timing of the transmitted signals is controlled by a closely speed-controlled motor. There are no pauses between the signal elements and the transitions between mark and space are as nearly instantaneous as the transit-time of the mechanisms and relays will allow. The signal may, therefore, be regarded for all practical puzposes as a square wave.

Reception of a start signal causes the selecting mechanism of the receiving teleprinter to be connected to a motor whose speed is regulated to within $0.5 \%$ of that of the sending motor. The purpose of the selecting mechanism is to investigate the position of the receiving relay at intervals corresponding to the

F.S.K. transmitter drive unit, Model PVT, made by Plessey.
middle of the five $20-\mathrm{msec}$. code elements to determine whether the incoming signals are mark or space. As each element is interrogated in turn the mechanism in the machine is progressively set until finally the type head is positioned and the letter is printed.

For reasons connected with the operation of the machine which we need not go into here, the receiving teleprinter needs 6 msec . of each code element to carry out its interrogation, thus, in the case of characters involving changes between mark and space, the change-over point cannot be advanced or retarded more than 7 msec . without danger of wrong selection and misprinting. Temporal displacement of the change-over points is called telegraph distortion and is expressed as the percentage of the $20-\mathrm{msec}$. element length by which the change-over is early or late. The maximum distortion the teleprinter will tolerate without danger of misprinting is $35 \%$.

## Waveform Distortion

Bias distortion is another phenomenon which may prove troublesome, and this arises when the receiving relay is not driven symmetrically in both directions by the incoming signal. Although bias distortion is not particularly serious when the incoming signals are perfect steep-sided square waves, if the pulses are rounded, sloped or otherwise distorted in transmission (as they may well be in a long radio link) the effect of bias distortion may be to alter the apparent time of change-over with consequent telegraph distortion. In " single-current " working such as on-off c.w. or single tone m.c.w., where a d.c. bias of half the signal level has to be introduced to polarize the receiving relay, not only is half the signal wasted but, in the case of varying signal strength, bias distortion arises if the signal departs from twice the bias value (as it may do with fading). Automatic bias control can mitigate this effect but has practical difficulties. F.S.K., being a truz "double-current" system in which signals are sent equally for mark as for space, has considerable advantages in this respect. It will further be seen that as distortion of the shape of the signal is detrimental any system should have as its aim preservation of the signal shape. F.S.K., being a c.w. doublecurrent system, furthers this end by making more efficient use of the transmitted signal than either c.w. on-off or single or two-tone m.c.w. It is further thought that certain effects of the f.m. nature of the signal have advantages in the presence of certain types of fading.

Having reviewed the method of reception of the radio signal and some of the problems connected with the teleprinter, we can now pass on to consider the
resolution of the teleprinter signal from the radio signal. Only the two main methods will be considered here, these being the linear and non-linear discriminator systems, usually referred to as the "discriminator" and "two-filter" methods.

In considering this phase of the operation we must bear in mind that we are conveying intelligence by varying a signal, whose frequency may be as high as $30 \mathrm{Mc} / \mathrm{s}$, by $850 \mathrm{c} / \mathrm{s}$ or less. Consequently, frequency stability is of more than secondary importance whatever system of reception is used.
In the discriminator system, Fig. 1, the r.f. signal is converted to a fairly low i.f. and passed through a limiter to a linear discriminator of the conventional f.m. type. The frequency-shifted signal will thus produce at the discriminator output a d.c. signal of positive or negative sign (polarized) which can be passed through a low-pass filter and further amplifiers, shaping circuits, etc., to the teleprinter. At first sight, it would appear that provided the discriminator be made with a sufficiently long linear characteristic, this system should be very tolerant of frequency drift, but in practice any drift causes a standing d.c. component to appear at the discriminator output resulting in bias distortion. Various methods of eliminating this bias have been evolved but they all involve other difficulties. The discriminator system is, however, much more tolerant of varying values of shift than the twofilter system and this is an advantage in some types of service as shift values are not yet by any means standardized.

The two-filter system, Fig. 2, involves the use of a heterodyne oscillator at i.f. to produce an audio note of 2,000 to $3,000 \mathrm{c} / \mathrm{s}$ whose frequency depends obviously on whether the transmitter is sending mark or space.

This audio signal is passed through a bandpass filter and limiters to two further filters tuned to the mark and space frequencies respectively. The outputs of these filters are rectified, filtered and dealt with as before to produce the polarized d.c. signal for the teleprinter. With the two-filter system in the case of small drifts the bias distortion per cycle of drift is very much lower than with the discriminator system. As long as the mark and space frequencies stay within the pass-bands of their respective filters there will be virtually no bias distortion. However, as these filters should for other considerations be made as narrow as possible it could be that the discriminator method would produce some signal from a carrier which had drifted too far to work with the two-filter system, but for many commercial applications, where accuracy is all-important, this might be a doubtful advantage.
fig. 1. Simplified diagram of a f.s.k. dual-diversity discriminator receiving system.



Fig. 2. Schematic diagram of a two-filter dual-diversity f.s.k. system.

It appears, therefore, that while the discriminator system may produce polarized signals under worse conditions than the other, the bias distortion in such circumstances would probably be excessive and at lower values of drift there is reason to believe that the twofilter system would produce lower distortion.

Work has been done in connection with both systems on the application of a.f.c. (automatic frequency control) and automatic bias adjustment at various points in the system but in the present congested condition of the short-wave bands, there is always a grave danger of the a.f.c. locking on to a strong interfering signal and causing a complete breakdown of communication.

It will be appreciated that in f.s.k. systems, the effects of level changes at the receiving aerial are less important than with some other systems since the continuous carrier simplifies a.g.c. problems and the ability to use a.f. or i.f. limiters further helps to reduce the effects of fading.

## Holding Signal

A further feature which has to be introduced into the telegraph converter arises from the arrangement whereby the sending teleprinter, when not sending traffic, transmits a continuous mark signal. This prevents the receiving machine from being operated by noise and also after a continuous mark has been received for a given time the motor of the receiving machine is stopped until a further start signal is received. Because an interruption of the carrier would result in the mark signal not being received, with a consequent chance of the receiving machine being started by noise, it is usual to incorporate in a radio teleprinter system a "mark-hold" device which automatically produces a mark signal locally at a fixed time after the last space pulse, in the absence of any further signal.

It is not proposed to deal in detail with transmitting systems as these were covered at some length in the article by Roddam already referred to. Two points are worth neting, however; first, that compared with the c.w. on-off system f.s.k. is much less likely to distort the signal elements in the transmitter. In the case of c.w. on-off (which is for all practical purposes a $100 \%$ square wave amplitude modulation) the keying, being usually carricd out in an early stage, may be subject to considerable distortion by the succeeding non-linear amplifiers. As there is a constant level of carrier in f.s.k., this trouble is avoided. The second point to be noted is that it is most important that the transition between mark and space be as smooth as possible; an abrupt and discontinuous change between mark and space will give rise to
spurious radiations and other undesirable effects.
Now to consider the question of frequency stability in relation to practical conditions. First, the comparatively small clear space allocated to each radio telegraph circuit makes it essential, no matter what system of telegraph conversion is used, to have the highest selectivity in the receiver compatible with conveying the intelligence. (In the case of an $850-\mathrm{c} / \mathrm{s}$ shift the bandwidth cannor safely be much less than $1,500 \mathrm{c} / \mathrm{s}$ ). The skirts of the i.f. response curve should be as steep as possible. With such selectivities the stability of both transmitter and receiver must be high and crystal control of the transmitter and of the receiver change-frequency oscillator is really essential if the system is not to fail by simple inability to convey any signal at all.

Stability of shift must next be considered. If the mark and space frequencies drift in opposite direction by the same amount no bias distortion will be introduced but if both drift in the same direction, or if only one drifts (both of which effects are more likely in practice) bias distortion will be intreduced, particularly in the discriminator system and in the twofilter case excessive drift of any type may overstep the pass-band of the filters.

It will be seen that a fairly high degree of frequency stability at all points in the system is essential, but if the station is to be continuously attended by a skilled operator who can devote most of his attention to a single radio channel it is possible (having made sure that neither shift, carrier nor receiver c.f.o. frequency are subject to violent and random changes) to provide monitoring devices and fine controls so that the operator may at all times keep the circuit at peak efficiency. A drawback of this system is that before a new circuit can be brought into service, some time must be wasted in lining up before traffic can be accepted as such a system cannot be adjusted in the absence of a signal from the corresponding transmitter.

## Unattended Operation

For certain types of service (such as aeronautical) the teleprinters may be located remotely from the radio station and many circuits may be in use at once under the control of operators whose accomplishments do not in any case lie in the direction of skilled adjustment of radio equipment. Similarly, it may be necessary to make many changes of frequency during a 24-hour period under circumstances which allow no time to be wasted in preliminary lining up. This problem has been the subject of some years' investigation by the writer's company in collaboration with International Aeradio and it is now belicved
that field test data has been collected on a scale sufficient to show that given stabilities of mean transmitted carrier, shift and receiver change-frequency oscillator and b.f.o. of the order of 1 part in $10^{5}$ (which is quite practicable) a system can be built which is capable of virtually unattended operation and in which new channels may be brought into use without preliminary lining up. A typical receiving equipment might have four dual-diversity channels and any one of four radio circuits can be selected remotely, but it is only necessary in this account to describe one channel.

A pair of receivers operating from spaced aerials has its c.f.o. signal provided by a separate unit using a temperature-controlled crystal oscillator in an oven stabilized by a temperature-sensitive resistance bridge.
B.F.O. signals to each receiver are supplied from another temperature-controlled unit which has alternative crystal-controlled and variable oscillators. The stability of all the above crystal oscillators is of the order of 1 part in $10^{16}$. Alarm circuits on all the ovens give warning should the temperature control fail.
A.F. signals from both receivers are passed to a converter unit via an input band-pass filter whence they pass through logarithmic amplifiers and limiters to the mark and space filters. The outputs from these filters are rectified and combined to produce a low-level polarized d.c. signal.

The converter unit also accomplishes the diversity switching. The signals from the receivers are made to operate an electronic switching device which selects the stronger of the two signals and suppresses the other. The characteristics of the switch are such that with the signals on both aerials subjected to a $20-\mathrm{db}$ fade the stronger signal is still selected as
long as there is a difference of 3 db . The output level of the converter is arranged to vary not more than 1 db for a $60-\mathrm{db}$ change in input.

The converter output is fed to a keyer amplifier unit where the signal is re-shaped and amplified to a level sufficient to work three teleprinters. The ability of the keyer amplifier to re-shape the signals to some extent is an important factor in operation under adverse conditions.

Various built-in test and measuring devices enable the apparatus to be set up in the absence of a signal. The variable b.f.o. facilities enable some degree of compensation to be carried out under local-control conditions when working against an unstable transmitter, but full realization of the benefits of the precise frequency control system necessitates the use of a very stable transmitting system and the drive equipment illustrated was designed to provide such a facility although there are many transmitters already in service of adequate performance.

The drive equipment uses h.f. oscillators and keyer amplifiers identical with those already described, together with shifter units giving crystal control of mean carrier and shift frequencies. These shifter units enable a number of preset channel frequencies to be selected by remote control and on each channel the shift is automatically adjusted to allow for the multiplication in use in the transmitter. A drive guard unit ensures that power cannot be applied to a transmitter unless drive is available. Warning devices operate should any of the ovens fail.

Using the two equipments described it is possible to set up a communications system on a number of frequencies between two points and to switch from one channel to another as propagation conditions dictate without any lining up or other attention and without interruption of the service.

# M(1)IELE IRAIDE(1) 

## Compact Sets Intouded for L'nskilled Operation

A RANGE of compact radio-telephone sets for use in motor cars, taxi cabs and various kinds of commercial vehicles and at the communicating fixed stations is now made by Hudson Electronic Devices, L.td., Appach Road, London, S.W.2.

The sets are intended for unskilled operation and ease of servicing, and work on fixed frequencies controlled by close-tolerance quartz crystals in the band 60 to $185 \mathrm{Mc} / \mathrm{s}$. While the same basic-circuit is used throughout, slight
differences are found in the sets for the higll and the low part of the band; for design purposes it is divided into two parts, low covering 60 to $100 \mathrm{Mc} / \mathrm{s}$ and high 101 to $185 \mathrm{Mc} / \mathrm{s}$. In general the high-band sets have one or two valves more than the low.

Transmitters follow well-tried practice, using a crystalcontrolled oscillator and frequency multipliers to give the working frequency and an r.f. power amplifier which is normally anode modulated. The r.f. output is about 5 watts for mobile and 6 watts for fixed sets.

Receivers are a little different to the usual in that they employ a double superheterodyne circuit with one oscillator only. This is crystal stabilized and different harmonics are selected for the two frequency changers.

Installation costs of such equipment must inevitably vary according to circumstances, but a typical mobile equipment complete works out at about $£ 85$ and a fixed station at $£ 120$, plus cost of aerial.

Alternative models are available for f.m. or a.m. and for rotary generator or vibrator operation.

Hudson v.h.f. mobile radio-telephone installation. The unit on the left is the power supply.


# Focusing Cathode Rays 

How Electron Lenses are<br>Arranged in C.R. Tubes

By "CATHODE RAY"

THIS is really a continuation of last month's issue ("Electron Optics") on the principles of electric deflection and focusing. So first will come a summary of the findings; then examples of how they are applied in cathode-ray tubes; and lastly something about magnetic focusing.

Here, then, is the summary. Cathode rays are streams of electrons in a vacuum. Each electron conforms very closely to Newton's laws of motion; that is to say, so long as it is not acted upon by any force it stays where it is or moves at constant speed in a straight line, and when a force acts on it in any direction it accelerates in that direction at a rate proportional to the force. An electron is so light that the force of gravity on it is negligible, but it responds smartly to electric and magnetic fields. An electric field is usually measured in volts per metre (or cm), and an electron placed in such a field accelerates positive-wards along the imaginary lines of electric force. These lines are at right angles to the equipotential lines, i.e., lines joining all points at the same potential. If the electric field is curved, or the clectron already has some velocity in a different direction, its track does not coincide with any line of force, but bends in towards it as the acceieration in that direction increases; just as a ball thrown out of a top-floor window does not immediately follow the lines of gravitational force downwards but curves gradually towards that direction. Consequently an electron's track can be bent more sharply by a given electric field when the electron is travelling slowly than when it is fast. Its speed depends on the field strength multiplied by the distance through which it has acted, and this product is the total potential difference; if it is denoted by V volts the speed reached by the electron from a standing start is $593 \sqrt{ } \mathrm{~V}$ kilometres per second. The track of an electron in a given electric field pattern can be calculated from these principles, but the practical problem, which is the reverse-to find the shapes and voltages of electrodes needed to produce a field pattern that will make the electrons in a cathode ray follow a desired pattern of tracks-is more difficult, and is largely a matter of cut and try. A help towards visualizing the relationship between electric field patterns and electron tracks is the analogy in which electrons are represented by little balls, electric field by gravity, intensity of field by gradient of the surface along which the balls roll, p.d. by difference in height, and equipotential lines by contour lines.

The problem in any cathode-ray tube is to persuade the electrons, which tend to fly off in all directions by mutual repulsion, to converge to a particular point somewhere on the fluorescent screen. If they were sent towards this point slowly, their mutual repulsion would again have time to work and they could not be crowded together to produce a sufficiently small and intense spot. But if they are shot towards it at many
thousands of miles per second they are there before they have time to realize that they hate one another's faces. So that is why a high voltage is applied between cathode and anode-which is sometimes very appropriately called the accelerator.

The converging, or focusing, can be done by either electric or magnetic fields, or of course both. In most oscilloscope tubes focusing is done by electric fields. In most television tubes it is done magnetically, but there is a tendency now to revert to electric focusing in order to save the cost of the magnet.
Before we tackle a complete c.r. tube it may help to clarify our mental pictures if we review the simple process of making a broad stream of marbles converge towards a point. Suppose they are being delivered uniformly along the front AB in Fig. 1 and that we want them to arrive at C. To get them moving in that direction it is of course necessary to make the surface slope down towards it. The speed they develop (friction neglected) is $8 \sqrt{ } \mathrm{H}$, where H is the total height in feet they lose in the process; this formula is analogous to $593, \mathrm{~V}$ for electrons. To obtain, in addition, convergence towards C the obvious method is to channel the slope into a sort of valley. The marbles respond best to this treatment near the

Fig. 1. Contour map of a valley slope
 (Fig. 2) for making marbles roll from $A B$ to $C$. The numbers attached to the con tour lines are in inches above $C$ level. The other dotted lines are the trocks of the marbles: note that by the time they hove reoched the last (zero) contour their momentum prevents them from rolling


Fig. 2. Model, of which Fig. 1 is the contour map. An electric field having equipotential lines of the same pattern as the contour lines of the model would make electrons follow similar tracks.
start, before they have got up enough speed for their momentum to have much effect. Once their tracks have been directed towards C , it doesn't matter if the rest of the ground is flat. One procedure would be to make a model as in Fig. 2 and vary the shape of the channel until the balls actually did all arrive at C. Fig. 1 could then be derived from it, as a contour map of the model, the numbers on the contour lines being "inches above C level." Note that at first the tracks are almost at right angles to the contour, but nearer the foot the marble's momentum (neglected in the similar diagram last month) makes them swing out slightly.

Alternatively, if one were better at maths than at modelling, the contour lines required to obtain the converging tracks could be calculated and plotted as in Fig. 1, and these contour lines used to form an experimental model as in Fig. 2, by which the calculation could be checked. If now an electrode system were designed to produce equipotential lines shaped exactly like the contour lines in Fig. 1, electrons released along AB ought to converge on C , provided that their speed was sufficient for their mutual repulsion to be negligible. Of course the potentials of the equipotential lines must be proportional to minus the heights of the contours.

Gravity models have actually been used in the practical development of electron lenses, but a more usual device is the electrolytic tank, mentioned last month.

## Typical Focusing System

Let us now follow the evolution of a typical electric focusing system or lens in a c.r. tube. To show what happens in the small holes through which the beam passes it is necessary to draw these parts larger than life, and if the whole tube were drawn on the same scale-well, the Editor would disapprove! So the diagrams that follow, which are longitudinal sections, are not accurately to scale.

The first requirement in the vacuum tube is a heated cathode, for emitting the electrons; and next, an anode for accelerating them in the desired direction. If the anode were simply a plate, the electrons would all crash into it and there would be no beam to light up the screen. So a hole is made in the centre of the anode, as in Fig. 3. Removing this bit of metal does not much affect the potential of the space it leaves, for that space is surrounded by metal at the


Fig. 3. Equipotential lines (dotted) and electron tracks in a tube having only a cathode and perforated anode. The space enclosed by the dotted boundary on the right is supposed to be all at the same potential as the anode.


Fig. 4. The addition of a cylindrical control electrode ("grid") at a slightly negative potential provides a more highly concentrated and controllable electron beam.
anode potential. So electrons starting off from the cathode scarcely notice the difference. As we saw last month, the equipotential lines are closer together where the lines of force diverge from a relatively small electrode, which means that the electrical slope is stecpest near the start, and the electrons accelerate very rapidly.

Electrons right on the axis go clean through the hole, for although the anode attraction becomes increasingly sideways as they approach it, the sideways pulls are equal all round and cancel out. Even electrons slightly off the axis are travelling so fast by the time they are near the hole that only those very close to the edge would be deflected sufficiently to hit the anode. In Fig. 3 a contour line has been drawn very close to the anode to show how the field tends to follow the shape of the anode into the hole. The tendency for the electron tracks to be attracted towards the lines of force at right-angles to this line is small, and does little more than cause the beam that goes through the hole to spread out slightly. To prevent it from being spread out more by positive surroundings beyond the anode, or being repelled back to the anode by negative surroundings, the space beyond the anode is kept at or near anode potential so that the electrons that get through continue on in straight lines.

An obvious disadvantage of the system so far is that most of the electrons go to waste by colliding with the anode, leaving only a feeble beam to light the screen. One step towards remedying this is to start the electrons off on the right lines by making them run the gauntlet of a negatively charged cylinder, as in Fig. 4. If this is made too negative it neutralizes the strong but distant attraction of the anode and prevents any electrons from getting out, so it serves the double purpose of beam-forming and controlling the amount of beam current, and by analogy with a valve (rather than any physical resemblance) is usually called the grid. One of its effects is to focus the beam at a point just beyond where the beam emerges from the grid. In a well-designed tube the diameter of the beam at this point (called the cross-over) is smaller than the emitting surface of the cathode, so in effect the electrons are coming from a close approximation to the ideal point source, to the benefit of subsequent focusing.
If the screen were just beyond the anode the spot of light would be little bigger than the hole in the anode. But to leave room for deflecting the beam over a reasonable area the screen has to be a considerable distance beyond the anode, and even if the hole


Fig. 5. Although the addition of another disk anode at a higher potential does not improve the focusing it is a step towords on effective electron lens-Fig. 6.
were extremely small the beam would spread too much to give fine definition, and, moreover (having come through such a small hole) it would be very weak. So the hole is made relatively large and the wide beam going through it is made to converge. The arrangement that does this is the electrons lens proper. There are innumerable varieties, but many of them comprise three anodes; in some, these anodes are kept at progressively higher potentials; in others, the first and third are highly positive and the second is less so, or even at or near cathode potential. Fig. 5 shows the sort of thing that would happen if the first and third anodes were disks with relatively large holes and the middle anode were omitted. The field between the anodes would be fairly uniform and would accelerate the beam but have little or no focusing effect.

Now put in a middle anode of cylindrical shape, as in Fig. 6. The metal wall of the cylinder short-circuits the electric field, for obviously it is all at the same potential, and so the equipotential lines are crowded together at the edges of the cylinder but are free to spread out in the centre. Remembering that the electrons tend to take the shortest cut from one equipotential line to another (but at the speed they are now doing they only tend to do so) we see that this pattern will make them converge. If it is found that the point to which they converge is, say, short of the screen, then by raising the potential of the middle anode some of the equipotential lines are transferred from the converging region to the diverging region, and the point of convergence is pushed farther away from the lens. This adjustment alone does not ensure that all rays converge on the same point to give a sharp focus; that depends to some extent on the voltages applied to the other anodes, but mainly on the shapes of all the anodes.

## Reversing Potential Gradient

A system basically like that just described is quite usual in oscilloscope tubes. The requirements for television are more exacting, and in a recent design there are four anodes: the first at only about +250 V , to attract the electrons gently through the grid orifice; the second and fourth forming a long cylinder with one section removed; and the third, between them, at about zero volts. The second and fourth are at about $+10,000 \mathrm{~V}$. One important thing to note about this and many other electric lenses is that the potential gradient reverses (in this example between the second and third anodes). After having fallen rapidly down a steep slope, the electrons have to go uphill
for a short distance, like a switchback. If they strike this upward slope head-on, at right angles to the equipotential lines, all that happens is that they are decelerated. The final velocity corresponds to the net difference of potential between start and finish. But if they strike the upward slope at an angle, that angle is increased by the slope (instead of being diminished, as it would be if the slope were downward). One can easily check this by rolling balls up a slope. Of course, if the slope rises to a greater height than the original starting point, it fails to clear the summit and rolls back; but in the electron lens with the zeropotential third anode this condition is avoided by placing it so that it is largely screened by the 10 kV anodes.

I hope that by the time of the next Radio Show someone will have made a gravitational model for the educational stand. A better method than messing around with plaster of Paris is to mount a sheet of rubber in a horizontal frame and clamp portions of it (corresponding to sections of electrodes) at heights representing the potentials of those electrodes. It could be a simple matter for the heights to be variable, showing the effects of focusing adjustments and also of deflecting potentials, about which I have said nothing, because their principle ought by now to be obvious.

Of course, it must be realized that although such a model is three-dimensional it represents the electric field pattern in only two dimensions-a cross-section of the electron lens, as in Figs. 1-6 here. The lens is (or should be) symmetrical around its axis, so that the same diagrams or models hold good for all longitudinal sections. Equipotentials are always surfaces; it is only in section that they appear as lines.

The reason why this three-dimensional situation can be discussed on two-dimensional paper is that the direction in which electrons are accelerated by an electric field is the same as the direction of the field (lines of force). And the reason why the effects of magnetic fields are so much more difficult to visualize is that the direction of acceleration is at right angles both to the direction of the electron's movement and the direction of the field, so even when the problem is presented in its simplest form it still involves three dimensions. And in a solemn session of the Institution of Electrical Engineers I once heard one learned gentleman after another confess that visualizing anything in three dimensions was quite a headache. (Viewers of "3D" films will doubtless heartily agree.) However, it is a difficulty that is basic to electro-magnetism. In


Fig. 6. An intermediate anode of cylindrical shape bends the equibotential lines (which are actually sections of equipotential surfaces) as here, making the beam converge to a point on the fluorescent screen.
ordinary electrical engineering-dynamos, motors, etc. -the problem is often a little easier because the electrons are confined to conductors, which are usually not free to move just anywhere. Electrons in a vacuum are not so bound. Their gyrations under the influence of combined electric and magnetic fields are the mathematicians' delight-but the non-mathematicians' despair.

However. Suppose we start with an electron gun something like that shown in Fig. 4, but with a larger hole, through which a slightly diverging beam of electrons is shot at high speed towards the screen. And suppose now that we wind a long coil around the beam, as in Fig. 7, producing a magnetic field pointing the same way as the electrons. What happens? To those electrons right at the centre of the beam, along its axis, nothing; for their path coincides exactly with the axial magnetic line of force, so they do not cut across the field, even slightly. So far as they are concerned there might not be a field. But those that are diverging do cut across the field, and so come under the law that makes electric motors motor. The appropriate memory-aider is the Left Hand Rule: if the thumb and first two fingers are stuck out all at right angles to one another, and the First finger is pointed in the direction of the magnetic Flux, and the seCond finger in the direction of the Current (which is opposite to the direction of the electrons) across the flux, the thuMb shows the direction in which the electrons tend to Move because of the electromagnetic force.

## Cathode's-eye View

Now in our c.r. tube the movement of the electrons, in so far as it is straight down the tube, paralle! to its axis, is not across the flux at all, so merely confuses the issue; the proper viewpoint for seeing the movement across the flux without the axial movement is from the cathode. So here, Fig. 8, is a cathode'seye view of the electron beam. Electrons along the axis remain at the centre all the time so do not appear to be moving at all; all others appear to be radiating outwards. Let us fix our attention on one particular electron just leaving the cross-over and diverging to the right. Since, for purposes of the left-hand rule, that is equivalent to a current in the opposite direction, the second finger should point to the left. The forefinger is pointing away, along the flux line, so the thumb shows that the force acting on the electron will make it accelerate downwards. If for simplicity we assumed (what in a real tube is not likely to be true) that the speed of divergence was constant, and also (what is certainly not true) that the direction of


Fig. 7. Section of electron gun, surrounded by a long coil designed to produce a nearly uniform magnetic field down the tube in the same direction as the electron beom.


Fig. 9. The divergent poths of electrons, os seen from the cothode, ore curled round into circles by the magnetic field. The time token to do o complete circle is the same for all electrons, slow or fost.

acceleration continued to be downwards, then the path of the electron would be a curve as shown dotted. This is the same kind of curve as that traced by a ball thrown horizontally, and for the same reason -that the ball is given a constant speed horizontally, combined with a steadily growing speed downwards.

In practice, during the first stage of the electron's flight, from cross-over to anode, it is not going at uniform speed but is being accelerated by an electric field; it is only from the anode onwards that its horizontal motion is at constant speed. However, this complication is largely offset by the fact that the electromagnetic force causing the electron to accelerate downward is proportional to the speed of the electron as well as to the strength of the magnetic field. The real complication is the second one-that the downward acceleration doesn't keep on being downward. It is always at right angles to the electron, so directly the electron starts curving downward as in Fig. 8 the magnetic acceleration veers round to the leftwards. This makes the electron curve all the quicker, which keeps the acceleration veering, and so on. When the ordinary mind tries to follow the electron it is therefore likely to become very dizzy. Even the chap who rather fancies his proficiency with the calculus and tackles the thing mathematically may quite possibly get himself into a mess.

But the situation is exactly similar to a very familiar one-a weight being whirled round at the end of a string, or, if you prefer, the earth revolving round the sun. The weight, let us suppose, is given a constant speed in a certain direction-say the original direction of the electron in Fig. 8. But being attached to the string it cannot go out in a straight line; it goes round in a circle centred at the point where the other end of the string is fixed. To keep it in this orbit, the string has to exert a tensional force. This force is obviously always at right angles to the direc-


Fig. 10. In practice a short magnet is used, and the effect on the electrons is more complicated.


Fig. 11. One effect of shortening the magnet is to change the circles of Fig. 9 into shopes more like this. The final result can still be satisfactory. and much more conveniently obtained.
tion the weight is going. And if the speed of the weight is constant, so too is the force. Since these are exactly the conditions that govern the movement of the electron, it is reasonable to suppose (and it can be confirmed mathematically) that the electron whirls round in a circle.

## Corkscrew Electron Tracks

All electrons have the same mass and charge, and we are assuming that the magnetic field is the same for all, so for a given speed they all experience the same force. Actually, of course, the speeds with which they cross the field are different; those only just off the axis diverge much more slowly than those on the outside of the beam. But if the equation for centrifugal force is applied to weights that are all the same, and force proportional to speed, it shows that the time for one complete revolution is the same for all, regardless of speed. The faster speeds are handicapped by having to take bigger circles. The same applies to electrons. So after a certain time all electrons that started off from the centre together, whether fast or slow, have done one revolution and arrive back on the axis again simultaneously. In the meantime, of course, they have been travelling down the tube, so the electron tracks are actually like a cork-screw-or a helix, to be more scientific. Fig. 9 shows one curl each of a few electron tracks as seen from the cathode. The small circles are made by slowly diverging electrons; the larger circles by the faster ones.

Increasing the strength of the magnetic field increases the centre-wards force on all the electrons, so tightens their curls and reduces the time they take to execute each complete turn. To get a well-focused spot it is obvious that the strength of the magnet must be adjusted so that one of the times when all the
electrons meet again is the moment they reach the fluorescent screen-which is a grand place for a reunion; they all get beautifully lit up.

Obvious it may be, but quite unpractical. To keep the electrons corkscrewing all the way to the screen it would be necessary to maintain the uniform magnetic field all the way there-a point that seems to have been overlooked in some of the drawings I have seen, purporting to explain magnetic focusing. Even to provide a magnet as long as the one in Fig. 7, extending as far as where the tube begins to open out, is too expensive and inconvenient, let alone one enveloping the entire tube, screen and all! So in practice a very short magnet is used, producing a curved field pattern something like Fig. 10. If you have been thinking that the theory of magnetic focusing has already been complicated enough, even with our beautifully uniform but quite unpractical field, you (and I) may well quail at the prospect of having to trace precisely what happens to diverging electrons in a field that varies rapidly both in strength and direction. But the operative word is "precisely." We can make a guess at roughly what happens.

At first, before an electron gets into the magnetic field, it is diverging quite happily in a straight line. But before it has had time to go too far, at the comparatively slow speed of this first stage of the journey, it finds itself going through the magnet ring, with a strong magnetic pull making it wheel round sharply. The magnet has been made of such a strength that by the time the electron is beyond its influence it has done an about-turn and is converging towards the axis again, as in the view from the cathode in Fig. 11. It is now also shooting towards the screen at really high speed, and, if everything has been done right, hits it just as it (and all its mates that left the cathode at the same moment) are on the axis again. Of course this is assuming the beam is not being deflected. If it is, then they meet elsewhere, but as all are equally deflected they do meet.
But don't ask me to produce a mechanical analogy to demonstrate all this! The proof of the thing is on your TV screen.

## Awarals ior Tectamical \athors

THE Radio Industry Council's premiums for technical writers for the year 1953 are now announced; as will be recalled, these awards are made with the object of encouraging the publication of clearly written expositions of British achievements in radio and electronics.

Premiums of 25 guineas each are awarded to the authors of the following articles:-
"Spectrum Equalization," by G. G. Gouriet (Wireless Engineer, May).
"Triode Transformation Groups," by A. W. Keen (Wireless Engineer, October).
"A Cylindrical Magnetron Ionization Gauge," by A. H. Beck and A. D. Brisbane (Vacuum, April).
"The Scanning Electron Microscope and the ElectronOptical Examination of Surfaces," by D. McMullan (Electronic Engineering, February).
"A Linear Sweep Cathode-Ray Polarograph," by H. M. Davis and Miss J. E. Seaborn (Electronic Engineering, August).
"Selective Calling for Radio-Telephone Systems," by J. R. Pollard (Electronic Engineering, December).

# Improved Radio Altimeter 

Sereo Principle Gizing Greater Freedom From Noise

By A. BLOCH.* K. E. BLECKS $\dagger$ and A. G. HEATON+

THE radio altimeter has now become quite a wellknown instrument for giving the pilot of an aircraft an indication of his clearance height above the terrain. Like many distance-measuring devices it works on the echo or radar principle, only instead of pulses it makes use of frequency modulation to obtain the actual measurement. A frequency-modulated continuous wave is transmitted downwards from the aircraft and is reflected back from the ground. By the time it reaches the aircraft again the transmitter frequency has changed, so there is a difference between the frequency of the received wave and that of the transmitted wave. This frequency difference is, of course, proportional to the time delay experienced by the returning wave, and so to the height of the aircraft above ground, and in the instrument it is obtained simply by heterodyning the two waves and taking the resultant difference frequency.

A new instrument has now been designed which works on this familiar principle but has improved sensitivity. It will give a reliable indication of height up to $5,000 \mathrm{ft}$ with an accuracy of $\pm 3, \ldots 5 \mathrm{ft}$ and over all types of ground-including dry desert which reffects only $3 \%$, of the incident power. At the same time the new instrument avoids certain complications, such as range-switching, which were necessary in previous designs. It uses a continuous-wave transmission, the frequency of which is varied linearly with time between 1,605 and $1,655 \mathrm{Mc} / \mathrm{s}$.

One method of obtaining the greater sensitivity has been by the use of the superheterodyne principle. This achieves the required large amplification of the received signal without introducing difficulties brought about by microphonic noise created in the amplifier by the vibration of the aircraft. The wave returned from the ground may have an amplitude of only one millionth or less of the amplitude of the wave transmitted and may thus require an amplification of $10^{\text { }}$ times in order to operate a robust indicator.
Instead of the returned wave being heterodyned with the transmitted wave, it is heterodyned with an auxiliary wave, the frequency of which is always $110 \mathrm{Mc} / \mathrm{s}$ higher than the transmitted wavc. This auxiliary frequency is created by mixing part of the output of the transmitting oscillator with the output of a $110-\mathrm{Mc} / \mathrm{s}$ oscillator and filtering out the desired component. If the returned wave had been heterodyned with the transmitted wave directly the resultant beat frequency would have been an audio frequency. Here, however, it is $110 \mathrm{Mc} / \mathrm{s}$ plus or minus this audio frequency, which is, of course, quite low compared with $110 \mathrm{Mc} / \mathrm{s}$, so that for brevity we can still speak

[^7]

Fig. 1. Simplified block schematic illustrating the principle of the altimeter.
Fig. 2. The cavity resonator of the transmitting oscillatnr, showing the small rotating cap:c.tor vane.

of an intermediate frequency of $110 \mathrm{Me} / \mathrm{s}$. This intermediate frequency is amplified about a thousand times and then mixed with the outpe: of the auxiliary $110-\mathrm{Mc} / \mathrm{s}$ oscillator. As a result, beats are produced of the same audio frequency as if the returned wave had been heterodyned directly with the transmitted wave. These beats are subsequently amplified and sent to an indicating mechanism, which will be described below. The exact frequency of the $110-\mathrm{Mc} / \mathrm{s}$ oscillator is immaterial, as long as it stays within the pass band of the i.f. amplifier.

Additional protection against noise is obtained by an application of the principles of servo mechanisms to keep the bandwidth of the amplifier smaller than it would be normally. In order to get a sufficiently accurate indication of heights below 900ft the constants of the altimeter are chosen so that for this height a beat frequency of just $10 \mathrm{kc} / \mathrm{s}$ is produced. Under the same conditions at $5,000 \mathrm{ft}$ a beat frequency of over $50 \mathrm{kc} / \mathrm{s}$ would be produced, so that an amplifier of this bandwidth would be required, which would pass five times as much noise as an amplifier with a bandwidth of only $10 \mathrm{kc} / \mathrm{s}$. In order to avoid this the following scheme has been adopted. The frequency variation of the transmitting oscillator is produced by a rotating capacitor vane driven by a small motor. Once a height of 900 ft has been reached, any further increase in height is arranged to cause a reduction of the voltage supplied to the driving motor, slowing down the motor until a beat frequency of $10 \mathrm{kc} / \mathrm{s}$ is restored. Obviously, the amount of slowing down of the motor is just as much a measure of the height reached as is the frequency attained with constant motor speed. The additional advantage of this scheme is that from a height of 900 ft upwards the beat frequency produced is always $10 \mathrm{kc} / \mathrm{s}$, and it is possible to make the audio amplifier most sensitive to this particular frequency. Thus the weaker signals, i.e., signals received at heights greater than 900ft, get maximum amplification.

## Circuit Arrangement

Fig. 1 shows the layout of the instrument and the relation of its various parts. The transmitting oscillator consists of a triode operating in a cavity resonator (Fig. 2). This resonator is modulated through the range $1,605-1,655 \mathrm{Mc} / \mathrm{s}$ by a small rotating capacitor vane driven by a special d.c. motor. (In this motor the speed of rotation is an accurate measure of the voltage applied to the terminals of the motor, a property required in the operation of the servo mechanism.) The larger part of the oscillator output is fed to the transmitting aerial. A small part is fed to the valve mixer, where it is mixed with the output of an auxiliary $110-\mathrm{Mc} / \mathrm{s}$ oscillator. This mixer valve, a triode, also operates in a cavity resonator which is of similar dimensions to the transmitting oscillator $c_{n}$ vity and is always kept in tune with the required frequency ( $110 \mathrm{Mc} / \mathrm{s}$ above the transmitter frequency) by a rotating capacitor vane mounted on the same shaft as the vane of the transmitting cavity. The output of this filter cavity is used as the "local oscillation" and is mixed with the received signal in the ring mixer.

The ring mixer is a so-called balanced mixer, i.e., a type in which two mixing elements (here silicon crystals) are operated simultaneously in such a way that the cffect of any amplitude modulation of the local oscillator frequency is cancelled in their com-


Fig. 3. Principle of the ring mixer. The local oscillation is fed to the crystals in phase while the returned signal is fed to them 180 out of phase. The i.f. outputs from the crystals are then in opposite phase and as they are combined by subtraction any unwanted amblitude modulation of the local oscillation is concelled out.


Fig. 4. The ring mixer consists of a coaxial transmission line, 1.5 wavelengths long, formed into a circle. This is an opened-up view.
bined output as far as possible (Figs. 3 and 4). The output of the ring mixer consists of a $110-\mathrm{Mc} / \mathrm{s}$ intermediate frequency with a superimposed audio beat corresponding to the height of the aircraft. This is passed through an i.f. amplifier, which has a bandwidth of $1.5 \mathrm{Mc} / \mathrm{s}$, and then on to a detector where the audio beat frequency is obtained. After amplification the audio beat goes to a counting circuit which produces an output voltage proportional to the rate of the beat.

It has already been mentioned that the altimeter has two different modes of operation: one, below 900 ft , where the modulating motor is run at constant speed and the frequency of the beats is taken as a measure of height; and another one, for heights above 900 ft , where the beat frequency is kept constant at $10 \mathrm{kc} / \mathrm{s}$ and the motor speed is taken as a measure of height. The altimeter must therefore include some mechanism which will automatically change from one mode of operation to the other one without an external switching operation.


One end of the chassis showing the ccuity resonator of the valve mixer and the capacitor-vane driving wheels.

The voltage from the counting circuit is compared with a voltage derived from a potentiometer that is mechanically coupled to the height indicator. Any unbalance detected is used to control the motor which sets the potentiometer and indicator in such a way as to eliminate the unbalance. For heights below 900 ft the voltage derived from the counter is (because of the constant speed of the modulating motor) proportional to the height of the aircraft. The voltage derived from the potentiometer is proportional to the height shown by the indicator. Thus the position of the indicator will always reflect the rate at which the beats are produced and therefore show the correct height (Fig. 5).

## Second Mode of Operation

The potentiometer is constructed in such a way, however, that once the indicator setting exceeds 900 ft it provides constant output, irrespective of height. Evidently its output can then only be balanced if the modulating motor is slowed down. For this purpose there is mechanically coupled to the first potentiometer a second potentiometer which supplies the voltage for the modulating motor. If the indicator shows a height of less than 900 ft this second potentiometer is in a "flat "portion of its range and thus supplies the required constant voltage for the modulating motor. But if the indicator goes to a height of more than 900 ft the potentiometer is in an appropriately graded part of its range and supplies to the motor a voltage which decreases with the height shown by the indicator. Hence, when the aircraft rises to a height of more than 900 ft the indicator motor-not being able to establish a balance on any setting of the indicator of less than 900 ft -will run on, drive the first potentiometer on to its "flat" portion and the second potentiometer on to its graded portion. It will thereby reduce the speed of the modulating motor until there is again


Fig. 5. Face of the height indicator. The full scole represents $1,000 \mathrm{ft}$, so the pointer makes five revolutions for $5,000 \mathrm{ft}$. Integral thousands are clockedup in the wincow at the bottom.
a balance between the output of the counter and the voltage supplied by the first potentiometer (now constant, corresponding to a beat frequency of $10 \mathrm{kc} / \mathrm{s}$ ). With proper grading of the second potentiometer the setting of the indicator at which balance is obtained and the indicator motor stops corresponds again to the height of the aircraft.

This second mode of operation continues for heights up to $5,000 \mathrm{ft}$. For heights exceeding this figure the motor speed stays constant and an indication is given that the aircraft is above the limiting height. This "hold off " presentation continues above $5,000 \mathrm{ft}$, even if the signal should fade into noise, as sufficient gain is provided before the counting circuit for this to be operated by noise. Such noise represents a signal above $10 \mathrm{kc} / \mathrm{s}$, since the audio amplifier characteristic extends beyond $20 \mathrm{kc} / \mathrm{s}$. It will be apparent that if an output corresponding to a frequency greater than $10 \mathrm{kc} / \mathrm{s}$ is produced by the counter, the indicator will run up beyond $5,000 \mathrm{ft}$ in an endeavour to find a balance point.

## Automatic Change of Characteristics

An additional feature of the altimeter is a system which indicates to the pilot that he is flying at a preselected height and warns him if he drops below this height.

An interesting point about the audio amplifier which magnifies the beat frequency is that it has a system for automatically changing its frequency characteristic to suit the mode of operation of the altimeter. Two input channels are provided. The first is in operation when the signal strength is below a certain level and its characteristic is of the peaked form referred to already. As the signal strength increases a bias is produced which closes this channel and the signal is now transferred to the second channel, which has a flatter characteristic. This arrangement has been chosen because when working at low altitudes; i.e., at low beat frequencies, a peak in the frequency response characteristic at $10 \mathrm{kc} / \mathrm{s}$ would cause an undesirable deterioration in the signal/noise ratio.

The aerials used are identical for receiver and transmitter. They are arranged on the aircraft in such a position that the direct transfer of energy from transmitter to receiver is kept at a minimum. An instailation on cither side of a tail plane is usually sufficient for this purpose. The directivity of these aerials is prescribed by operational requirements such as angles of bank, dive, or climb during which the reflected signal must be received.

# Attenuators $\operatorname{lor}$ High 

## Frequencies

By R. F. PRIVETV', ${ }^{\star}$ m.s.

Basis for a Design Using Standard Components

IIANY r.f. measurements, such as the measurement of receiver gain, require that a known voltage be produced at the input terminals of the unit under test. The input voltage may be too low to permit it to be measured directly, and it is then usual to generate the signal at a level convenient for measurement, to attenuate it to the required level and to transmit it via a screened conductor to the unit under test The combination of oscillator, monitor, attenuator and coaxial cable is called a standard signal generator. This article is intended to give the reader an understanding of that part of a signal generator which is in general the least understood-the output system.
"Cathode Ray" has given an introduction to the subject in which he stresses the importance of matching and illustrates "ladder" attenuators and attenuators which contain "T-section " pads of fixed attenuation, which are selected by the operation of two-pole, two-way, changeover switches. The latter system is the one described in this article, and, as designed, the attenuator can provide any attenuation variable in increments of 1 db , up to a total of 80 db , by the combination of pads of $1,2,5$ and 10 db loss.

To appreciate the importance of each factor involved in attenuator design it is necessary to consider the following topics: (1) Precautions required in the transmission of r.f. energy and in particular propagation along coaxial cables. (2) The r.f. behaviour of resistors, and the variation of pad attenuation with changes of resistance. (3) The cffects of switch capacitance and lead inductance.

In a signal generator one must also consider how best to measure the attenuator input.

## The R.F. Transmission Circuit

An output system is illustrated in Fig. 1. The open circuit voltage E of the generator may be measured and held constant and the voltage presented to the unit under test is the voltage $e$ developed across the cable termination $Z_{T}$. This diagram has been simplified and shows the generator as a source of opencircuit voltage $E$ and internal impedance $Z_{*}$, the attenuator as comprising one " T -section" and the termination as an impedance $Z_{T}$ separate from the unit under test (in practice this unit may provide a part, or all of this terminating impedance).

We will now concern oursclves solcly with the requirements placed by the cable on $Z_{T}$ and $Z_{8}$. The physical form of the generator and the behaviour of the attenuator will be discussed later. So let us

[^8]connect the generator directly to the cable by putting $\mathbf{R}_{1}=\mathbf{R}_{3} \quad 0$ and $\mathbf{R}_{\mathbf{2}}=\infty$.
The characteristic impedance $Z_{w}$, of a cable may be defined as the input impedance presented by an infinite length. A lossless cable has a $Z_{0}$ which is purely resistive and can be calculated in terms of the geometry of the conductors and the dielectric constant of the insulation. Conductor resistance and dielectric loss cause $Z_{\text {ı }}$ to have a reactive component and cause attenuation along the length of the cable, which, however, can be minimized by the choice of an optimum value for $Z_{0}$. The standard polytheneinsulated cable is designed for this optimum and has a $Z_{0}$ which may be taken as resistive, and is nominally 75 ohms with a manufacturing tolerance of a few ohms. For our purpose the cable will be considered lossless.
A length of cable can nevertheless affect the output by introducing impedance transfurmations; for example a cable of length one quarter of a wavelength $\left(l=\frac{\lambda}{4}\right)$ can behave as a transformer causing our generator of impedance $Z_{s}$ to appear at the other end of the cable as a generator of impedance $Z^{\prime}=\frac{Z^{2}}{Z^{2}}$. The open circuit voltage E would be transformed as the square root of the impedance, i.e. $E^{\prime}=E \sqrt{\frac{Z^{\prime}}{Z}}$. From the above it follows that the variations of the output voltage $e$ due to changes with frequency in the electrical length of the cable (defined as $\frac{2 \pi l}{\lambda}$ ) can be reduced to zero by making $Z_{\mathrm{s}}=Z_{6}$. The generator is then matched to the cable and the value of the load $\mathrm{Z}_{\mathrm{T}}$ is unimportant. By a similar process one may show that the generator impedance $Z$, is unimportant if $Z_{T}=Z_{0}$.


Fig. 1. Simplified circuit of the output system of a signal generator.

In practice the cable may not be uniform and the source or generator may not be perfectly matched. Then it is best to have the cable approximately matched at both ends.

It is possible to design the generator impedance to match the cable, and imperfections are more likely to be present at the termination. Part or all of the terminating impedance may be provided by the unit under test, and the effects of mismatches introduced by the unit must be minimized.

As an example of the use of a matched generator in minimizing the effects, consider the problem of providing a constant voltage at the input terminals of an amplifier which has a small input capacitance $C$. If these terminals are shunted by a resistor the cable can be terminated correctly at low frequencies, but the capacitance will introduce a moderate mismatch at high frequencies. The voltage developed across the terminals has been calculated, by methods outlined in the Appendix, for the case of a matched generator, and for that of a zero impedance or constant-voltage generator. The results for the particular values of $\mathrm{C}=10 \mathrm{pF}$, cable $\mathrm{Z}_{0}=75 \mathrm{ohms}$ and length $l=$ 1 metre are plotted on Fig. 2. The dotted curve, calculated for a generator of 75 ohms output impedance, shows a loss increasing with frequency to a value of 0.86 db at $200 \mathrm{Mc} / \mathrm{s}$, whereas the full curve, calculated for a zero impedance source, shows large fluctuations with frequency; even at $30 \mathrm{Mc} / \mathrm{s}$ the output has risen by 0.5 db .

We may conclude that a generator of the correct output impedance is essential, unless measurement is restricted to frequencies at which the cable has a length less than one-eighth of a wavelength, or to frequencies at which the unit under test can be assumed to match the impedance of the cable.
R.F. Behaviour of Resistors and Resistance Networks.-The low-frequency accuracy of an attenuator using resistance networks is limited by the stability of the component resistors. D.c. measurements of resistance and successive adjustments may be employed to approximate to the required value of resistance with any accuracy likely to be desired, until the error remaining is of the order of the errors expected to arise with age, temperature, humidity, etc.

Unfortunately the resistors of the higher stability generally become extremely reactive at high frequencies, and attenuator design has shown signs of division into two classes. The first class is exemplified by the precision instrument accurate to a hundredth of one db which contains wirewound resistors, and the second by the instrument of medium d.c. accuracy which is maintained at high frequencies. Research is in progress to improve the highfrequency performance of the former class and the accuracy of the latter. The attenuator described falls into the latter class and employs the "cracked"
carbon high-stability type of resistor. As a useful attenuator can be constructed using the familiar resinbonded carbon type of resistor both types will be considered.

The resistance values required for atrenuator pads of 75 ohms characteristic impedance are given in Table I. Each pad is symmetrical so that, in the nomenclature of Fig. $1, \mathrm{R}_{1}=\mathrm{R}_{3}$.

A pad in which all the resistances are the same has attractions economically and, as will be shown later, electrically. This pad gives an attenuation of 11.44 db . The values are calculated from the formulx :-

$$
\begin{aligned}
& R_{1}=Z_{0}\left(\frac{N-1}{N-i} 1\right) \\
& R_{2}=Z_{0}\left(\frac{2 N}{N^{2}-1}\right)
\end{aligned}
$$

where N is the ratio of input current to output current, which in the case of a 3 db pad is $\sqrt{ } 2$.

The resin-bonded carbon resistor can be obtained in miniature form with ample dissipation for attenuator inputs of a few volts r.m.s., and the required value of resistor can be approximated to within $\pm 5$ per cent by sclection from the preferred standard values of the 5 -per cent tolerance range.

For example, half the resistors nominally 20 ohms or 22 ohms should be within 5 per cent of the 21 -ohm value required by the 5 db pad. With a temperature coefficient averaging 0.05 per cent per degree $C$ and a shelf-life stability of $\pm 2$ per cent, selection to $\pm 3$ per cent would be profitable.

The errors in attenuation between correct values of source and load impedance due to pad resistances of 90 per cent of the correct values are given in Table 11 .

The values in brackets are those obtained with a


Fig. 2. Effect of source impedance on output voltoge from a I-metre cooxial cable feeding a resistonce equal to the characteristic impedance shunted by 10 pF .
zero-impedance source and it can be seen that a correct source impedance is important in reducing the effect of errors. The change in attenuation due to a proportionate change in all resistances is then low, and this has a bearing on high-frequency response (as will be shown).
"High-stability" resistors, stable to $\pm 1$ per cent,


Fig. 3. Equivalent circuit ofo carbon resistor.


Right: Fig. 4. Underside of attenuator using wafer switches. Foil instecd of wire is used for cross leads to reduce inductance.

TABLE 1

| Attenua- <br> tion (db) | 20 | 11.44 | 10 | 5 | 3 | 2 | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $R_{1}$ | $\ldots$ | 6136 | 43.30 | 38.96 | 21.01 | 12.8 | 8.66 |
| $R_{2}$ | $\ldots$ | 15.15 | 43.30 | 52.71 | 123.4 | 212 | 323 |

TABLE 11

| Nominal attenuation (db) | Increase in attenuation (db) due to - 10 per cent resistance change |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | $\mathbf{R}_{1}$ only | $\mathbf{R}_{1}$ and $\mathbf{R}_{3}$ | $\mathrm{R}_{2}$ only | $\mathbf{R}_{1}, \mathbf{R}_{2}$ and $R$ |
| 11.44 20 | -0.05 -0.26 -0.38 | $\begin{array}{r} -0.15 \\ -0.51 \\ (-0.8) \\ -0.72 \end{array}$ | $\begin{array}{r} +0.17 \\ +0.54 \\ (-0.4) \\ +0.76 \end{array}$ | $\begin{aligned} & +0.02 \\ & +0.02 \\ & (-0.28) \\ & +0.02 \end{aligned}$ |




Fig. 5. Equivalent circuit of switch and wiring and variation of losses with frequency.
can be obtained with a nominal dissipation of ${ }_{x}^{1} \mathrm{~W}$. They are slightly larger than the conventional resistors described, but in spite of this, and the spiralling of the resistive track which is necessary to control the value of resistance, they have high-frequency properties which are very similar. An equivalent circuit which approximates to both types of resistor in the resistance range of 10 to $1,000 \Omega 2$ is shown in Fig. 3.

The shunt capacitance $C$ is usually below 1 pF and the series inductance will vary with lead length, but should be under $0.02 \mu \mathrm{H}$. It thus appears that the behaviour of a resistor will vary with its resistance value $R$ and it is found that the impedance of a 1,000- $\$ 2$ resistor will fall at high frequency and that of a $10-1!$ resistor will rise. Resistors in the range 40 to $200 \Omega$ can be expected to retain a resistance close to their d.c. value.

Reference to Tables I and II will now show why it is advisable to avoid low values of resistance, especially when included in a pad of high attenuation, e.g. the 15- $\$ 2$ resistor in the $20-\mathrm{db}$ pad. The advantages of using pads with equal resistors, having the same behaviour of high frequencies, are now obvious. This suggests that the bulk of the attenuation should be provided by pads of 11.44 db attenuation and to approximate to this value it was decided to standardize on a $10-\mathrm{db}$ pad.

Switches and Interconnections.- Two-pole twoway switches are required to control the insertion of the pads and many varieties of toggle and wafer switches are available. The wafer type has the lowest capacitances to earth and across the contacts, but its use leads to a large attenuator compartment. Fig. 4 gives an underneath view of an attenuator using wafer switches and containing 20, 10 and 5 db pads, giving a total attenuation of 75 db variable in increments of 5 db . Such a design using pads of 10 db maximum attenuation and giving increments of one db would result in a very bulky unit, but it is possible that 20 db pads in which the $15-\Omega$ resistor is replaced by 3 resistors each of $45 \Omega$, would be satisfactory when used in conjunction with wafer switches. Space could be saved by using toggle switches to control the insertion of the pads of lower attenuation.

Some attention was given to the performance of the attenuator at its zero attenuation setting, as heavy losses due to the switches and connecting leads would render the attenuator useless for incorporation in a signal generator. Measurements were made of the input impedance $Z_{i}$ of the wafer-switch type of attenuator at its zero setting with a $75-\Omega 2$ load across the output terminals. The results of these measurements together with the equivalent circuit of a switch and its connections are shown in Fig. 5.

The reduction of the inductance $L$ of the connection across the switches, which was achieved by replacing


Fig. 6. Attenuator unit for frequencies up to $100 \mathrm{Mc} s$ using standard toggle switches.
the original copper wires by strips of copper foil, resulted in an input impedance which varied slowly with frequency. These strips can be seen in Fig. 4 and the theoretical explanation of this behaviour is that the series inductances and shunt capacitances form a low-pass filter which by the reduction of inductance has been brought to a characteristic impedance, analogous to that of a cable, of approximately $75 \Omega$. The cut-off frequency of this filter is of the order of $500 \mathrm{Mc} / \mathrm{s}$ so that at frequencies up to $100 \mathrm{Mc} / \mathrm{s}$ the attenuator at zero setting
behaves very much as a length of $75-\Omega$ cable.
With the application of these principles attenuators were constructed using "Arrow," "Bulgin" and "N.S.F." toggle switches, and the type produced in small quantities is that illustrated in Fig. 6. The resistors used were supplied by special arrangement to be within $\pm 1$ per cent of the required values and are high-stability components. Standard P.O. type connectors were used while the box was fabricated from sheet brass.

One of the units was sent to the National Physical

## APPENDIX

A generator of impedance $Z_{s}$ is connected by a cable of characteristic impedance $Z_{n}$, and length $l$, to a termination consisting of a resistance $Z_{0}$ shunted by a capacitance $C$. If the generator impedance $Z_{0}=Z_{n}$ the output voltage $e$ developed across the termination will be independent of $l$ and the circuit has the simple equivalent form shown in Fig. Al.

We then have $\frac{e}{e_{0}}=-\frac{1}{1+j \overline{\omega C} \overline{Z_{0}} 2}$ and taking the modulus or "amplitude of $e$ "

$$
\left|-\frac{e}{e_{0}}\right|=\frac{2}{\sqrt{ }_{\prime}^{\overline{4+\omega^{2}}}=\overline{\mathrm{C}^{2}} \bar{Z}_{01}}
$$

If the generator impedance $Z_{s}=0$ we have the equivalent circuit of Fig. A2.
The equivalent output impedance is reactive and the open-circuit voltage is transformed in the ratio $\sqrt{1+y^{2}} \div 1$
where $y=\tan \frac{2 \pi l}{\lambda}$. The term $\frac{2 \pi l}{\lambda}$ is called the electrical length of the cable and the wavelength in the cable is $\lambda=, \frac{\lambda_{0}}{\kappa}$ where $\lambda_{0}$ is the wavelength in air and $\kappa$ is the dielectric constant of the insulation. (For polythene $\kappa=$ 2.23 and $\lambda \bumpeq \frac{2 \lambda_{11}}{3}$ )
lirom the above circuit

$$
\left|\frac{e}{e_{0}}\right|=-=\frac{V^{\prime} 1+y^{2}}{\sqrt{\left(1-y Z_{0} w C\right)^{2}+y^{2}}}
$$

When $y=0\left(l=\begin{array}{c}n \lambda \\ 2\end{array}\right) \frac{e}{e_{0}}=1$, and for a cable of length one metre this occurs at $0,100,200,300 \mathrm{Mc}$ 's etc. Note that $e=e_{0}$ at all frequencies if $l=0$, and that $e=e_{0}$ for all $l$ if $\mathrm{C}=0$.



Fig. 7. Typical coupling circuit between oscillator and attenuator.

Laboratory for calibration at high frequencies and the results are summarized in Table III. The change in attenuation relative to the value measured at $1 \mathrm{Mc} / \mathrm{s}$ is given for various combinations of pads at frequencies of $25,50,75$ and 100 Mc 's. At the higher frequencies the measurements were limited to a maximum attenuation of 50 db . Also the attenuation introduced by the presence of the attenuator at its zero setting was measured.

Measurement of Attenuator Input.-Consider the circuit coupling an oscillator to the input of the attenuator and in particular the arrangement in Fig. 7. Here a pick-up coil, which is lightly coupled to the oscillator, and is damped by resistance $\mathrm{R}_{1}$, develops a voltage across a diode voltmeter. It is normal practice

TABLE III

| Pads in circuit | Nominal attenuation (db) | Attenuation in db relative to $1-\mathrm{Mc} / \mathrm{s}$ value at frequencies of : |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\mathbf{2 5 ~ M c / s}$ | $50 \mathrm{Mc} / \mathrm{s}$ | $\begin{gathered} 75 \\ \mathrm{Mc} / \mathrm{s} \end{gathered}$ | $\begin{gathered} 100 \\ \mathrm{Mc} / \mathrm{s} \end{gathered}$ |
| $1 \times 10 \ldots$ | 10 | 0 | 0.05 | 0.025 | 0.05 |
| $1+2+2+5$ | 10 | 0 | 0.05 | 0.10 | 0.125 |
| $5 \times 10 \ldots$ | 50 | 0 | 0.10 | 0.25 | 0.10 |
| $\begin{array}{r} 1+2+2+5+ \\ (4 \times 10) \quad \cdots \\ \text { All } \end{array}$ | 50 80 | -0.05 -0.15 | +0.075 -0.125 | 0.3 | 0.2 |
| "Zero level" insertion loss | 0 | 0.15 | 0.25 | 0.45 | 0.65 |

to hold this voltage to a constant value at all frequencies, either by an operator who corrects errors shown on a meter by the adjustment of manual controls, or by a feedback network.

Whether the servo system is automatic or is completed by human agency it results in a constant voltage $E$ being maintained across $R_{1}$. At the attenuator input the terminals yield an open-circuit voltage E and a short-circuit current $\mathrm{E} / \mathrm{R}_{2}$ and by Thévenin's theorem must appear as a source of resistance $\mathbf{R}_{2}$. Thus to suit the attenuator described $R_{2}$ would be made 75 s. A low value of $R_{1}$ will minimize the adjustment necessary to compensate for changes of load presented by the attenuator, but too low a value will cause excessive loading of the oscillator circuit.

#  

Tro Programmes in a Standard Channel

With frequency modulation very much in the air as a result of the Television Advisory Committec's recent report on v.h.f. broadcasting, it is interesting to hear from America about a system of f.m. multiplexing whereby a standard $200-\mathrm{kc} / \mathrm{s}$ channel can be used for carrying two programmes simultancously. It is the work of the late Maj. E. H. Armstrong, the pioneer of f.m. transmission, and J. Bose of Columbia University, and has been undergoing tests for several years at Maj. Armstrong's experimental f.m. station in New Jersey. The auxiliary channel is provided by a $27.5-\mathrm{kc} / \mathrm{s}$ sub-carrier, which frequency modulates the main carrier and is itself frequency modulated by the auxiliary programme.

Previous attempts have been made at f.m. multiplexing for broadcasting but have not been very successful, mainly because of cross-modulation between the main and auxiliary channels and the transfer of noise from one to the other. Armstrong and Bose have published claims to have overcome this trouble by arranging for the modulation processes of the two channels to be performed in separate parts of the transmission system, so that they are protected from each other. The phase-shift method of modulation is used, with frequency multipliers to increase the small deviation obtained from it to the required value. A crystal oscillator providing the drive is followed by a phase-shift modulator into which is fed the a.f. modulating signal of the main programme.

The resulting modulated carrier then passes through a chain of three frequency multipliers and interposed band-pass filters to a power amplifier and the aerial, where the final maximum deviation is about $50 \mathrm{kc} / \mathrm{s}$. Between the second and third multipliers is inserted a second phase-shift modulator and here the $27.5-\mathrm{kc} / \mathrm{s}$ sub-carrier with its auxiliary programme is superimposed on the main carrier. The deviation produced by this sub-carrier is of the order of $\pm 20 \mathrm{kc} / \mathrm{s}$, while its own deviation by the modulation is about $\pm 5 \mathrm{kc} / \mathrm{s}$.

The phase-shift modulation system is also used within the auxiliary channel for impressing the audio input on the sub-carrier. To obtain the necessary deviation, however, the frequency multipliers have to increase the modulated sub-carrier to something like $11 \mathrm{Mc} / \mathrm{s}$, so a frequency converter is used to bring this down to the required $27.5 \mathrm{kc} / \mathrm{s}$.

Demonstrations of the system have been given with different programmes on the two channels, and in spite of the fact that the modulation in the auxiliary channel is limited to $7.5 \mathrm{kc} / \mathrm{s}$, compared with the main channel's $15 \mathrm{kc} / \mathrm{s}$, listeners at the receiving end have reported "no audible difference in quality between the two programmes." Other demonstrations have been given with the two channels carrying the same programme; with binaural audio inputs; and with the receiving loudspeaker listening to the silent auxiliary channel to show the absence of crossmodulation from the main channel.

# Vibratory Gigroseope 

## Electronic Indicating System

THE familiar tuning-fork, once used for controlling the frequency of wireless transmitters, now figures as the stabilizing element in a new kind of gyroscope devised by the Sperry company. It performs the same function as the spinning flywheel of an ordinary gyroscope, but in rather a different way. Actually it operates on much the same principle as the physiological balancing mechanism of the common housefly, which has club-shaped vibrating rods situated just behind the wings. The main advantage of this type of instrument is, of course, that it has no rotating parts or bearings to cause trouble. In addition, its sensitivity to different rates of turn is claimed to cover a very wide range-from as slow as the earth's rotation to as fast as 100 r.p.m.

In operation the two prongs of the tuning fork (Fig. 1) can be regarded as segments of two ordinary gyro flywheels pivoted at somewhere near the heel of the fork. Since the prongs are vibrating towards and away from each other, the imaginary flywheels would be continually reversing their direction of rotation in a similar manner, one going clockwise to the other's anti-clockwise. If the "flywheels" were rotating continuously as in an ordinary gyroscope they would provide a steady reaction against any external force that tried to turn the tuning fork about its vertical centre line. Since, however, their motion is actually alternating, the reaction to an applied turning force is also alternating, and it appears as a vibratory twisting motion in the shaft of
 the fork. The greater the rate of turn applied to the fork the greater the amplitude of this torsional vibration. As for the direction of turn, clockwise or anticlockwise, this is indicated by the phase of the torsional vibration. With clockwise turning, an inward swing of the vibrating prongs causes the torsion blade to approach the lefthand pick-up coil, while with anti-clockwise turning an inward swing sends it

Fig. 2. Cut-away model showing the actual form of the instrument. Underneath the prongs is the tuned torsion shaft, and at the base of this ore the torsion pick-up coils.


Fig. I. Essentials of the experimental vibratory gyroscofe.
towards the right-hand coil. In the actual instrument the shaft of the fork is mechanically tuned so that it resonates torsionally at the natural vibration frequency.

As can be seen from the diagram, the prongs of the fork are kept in vibration (actually at $1,850 \mathrm{c} / \mathrm{s}$ ) by a feedback arrangement comprising driving coils, pickup coils and an amplifier. The resultant torsional vibration is detected by a blade on the shaft and two more pick-up coils. The electrical signal from these is amplified and applied to a phase detector, which also has an input signal from the vibrating prongs to provide a reference phase. Thus the phase detector gives the phase of the torsional vibration relative to the prong vibration and hence the direction of turn, as already explained. Its output is in the form of a d.c. signal which indicates by its polarity the direction of turn and by its amplitude the rate of turn. These indications are shown by appropriate pointer movements on a centre-zero meter.

## P(DWERTMINNSENTIIR

ACCORDING to a description in the December, 1953, Electronics a 20 -watt transistor, which exceeds the normal power level by about 100 times, is being produced by the Minneapolis-Honeywell Regulator Company. It is a diffused-junction germanium transistor hermetically sealed in glass and metal and having a screw mounting for direct connection to a metal chassis, or other support, which will act as a conduction cooling medium.

For the present they are being used in aircraft fuel gauges only but some may become available for commercial use later. No mention is made of their frequency characteristics but a 20 -watt transistor excites the imagination and conjures up visions of economical and compact power amplifiers.

## Ionospliere IReview-lisidis

## A Correction

THE two graphs on page 67 in last month's issue were unfortunately transposed. The lower graph is Fig. 1 (twelve-month running averages of sunspot numbers, etc.), while the upper is Fig. 2 (monthly mean sunspot numbers, etc.).

## Mobile Flight Simulator

## Self-contained Unit

## for Training Canadian

Cut-away model showing layout of the mobile flight simulator. A, air-conditioning plant : B, power supply ; C, computor rocks ; D, instructor's console ; E, cockpit.

THE first of the ten fligint simulators ordered by the Canadian Department of Defence Production was handed over at a ceremony in London recently by the designers and makers, Redifon, Lid., Broomhill Road, London, S.W. 18 .

This equipment is for the training of "Sabre" jet fighter pilots, and can reproduce all the effects of flight on instrument readings and the "feel" of the controls, with appropriate sound effects. In everything but the accelerations resulting from change of course and speed the pilor feels that he is airborne, and can be set problems in navigation, artificial emergencies due to such causes as icing, engine defects, etc., any of which might have serious consequences for an inexperienced trainee if he were in artual flight. The instructor can watch the pupil's reactions and correct errors in safety, and, last but not least, at low cost. Taking everything into account the simulator costs about $£ 3$ per hour to run, whereas the overall cost of flight training would be $£ 50$ per hour. Not the least of the advantages of flight simulators is the saving of training time which might otherwise be lost through bad flying weather. They can also be used to re-examine qualified pilots periodically, and tests can be devised which are far more searching

On the right are shown recordeps for plotting simulated flight poth. These also generate long-ronge d.f. signals.

One of the computor-amplifer racks is shown in the photograph on the extreme right. Six of these are mounted "herring bone" fashion in the middle section of the von.


# Manuiacturers’ Products 

NEW F(OUIPMENT ANI) ACCESSORIES FOR RADIO AND ELECTRONICS

## Tuner Unit

THE detector circuit in the "Elpico" Model RF/716 superheterodyne tuner unit makes use of a double triode, the first half of which

" Elpico " Model RF; 716 tuner unit.
functions as a diode and the second as a cathode follower. By this means capacitive couplings and shunts have been eliminated and the d.c. and a.c. load conditions are the same. It is claimed that the circuit will accept, without adding distortion, the maximum modulation level used by the B.B.C. Diodes for a.g.c. are combined with the pentode i.f. valve.

Long, medium and short wavelengths are covered and the unit, which is made by Lee Products, 63, Great Eastern Street, London, E.C.2, costs £14 14s.

## Sapphire Needles

THE installation of automatic grinding and polishing machines has enabled the makers of "Windsor" sapphire-tipped needles to produce a range retailing at 2 s 6 d each. These are all of 0.0025 in tip radius for 78 r.p.m. records and are available" in standard diameter or " midget" shanks, or in the "trailer" type for use with older heavy magnetic pickups.
The makers, Sapphire Bearings, Ltd., 16, Catherine Place, London, S.W.1, state that every needle is inspected by shadowgraph for sphericity and radius of point, and surface polish.

## Miniature Trimmers

COMPRESSION capacitors of the " postage stamp" variety for circuit trimming are now obtainable from the British Distributing Company, 591, Green Lanes, London, N.8, in capacitances ranging from $5 / 60 \mathrm{pF}$ ( $\min$ and $\max$ ) to $150 / 400 \mathrm{pF}$. They are assembled on ceramic bases designed for use separately or in banks
of any number and mixture of capacitances. The dielectric is mica. Individual units measure $\frac{3}{4} \times \frac{3}{6} \times \frac{2}{3}$ in and the test voltage is 300 d.c.

## Germanium Crvstal Coil

AMONG the small r.f. coils made by The Teletron Company, 266, Nightingale Road, Edmonton, London, N.9, is one designed especially for a germanium crystal receiver. Described as the Type HAX it has three separate windings on a 0.45 -in moulded former with an adjustable dust-iron core. The tuned winding is Litz-wound and covers the medium broadcast waveband with a $0.0003-\mathrm{mfd}$ variable capacitor.

The other windings are for aerial and crystal respectively, the couplings being set for optimum performance with an average aerial. The coil costs 3 s .

## Potted Transformers

A HARD-SETTING resin compound is used for casing a new range of miniature audio transformers introduced by John Bell and Croyden, 117, High Street, Oxford. With this type of finish fixing screws can be moulded into the block where required and become an integral part of the component. Rigid tags or pins can replace loose connecting wires and be inserted where required for particular purposes. Non-standard shapes and sizes are more easily provided by potting than by most other types of finish.

## Electrostatic "Tuceter"

THE introduction of new materials and methods has brought about a revival of interest in the electrostatic loudspeaker, particularly in Germany. The two outstanding advantages of this type are the simplicity of application (it is connected across the output transformer primary with a simple filter to keep out low frequencies), and the virtual absence of directional properties through 180 degrees in the horizontal plane.
One example, the Körting, is now available in this country through
E. D. Parchment, 69, Clapham Road, London, S.W.9. It consists of a curved perforated back-plate with a stretched diaphram of plastic film, coated on the outside with a metallic film. The whole is housed in a moulded plastic frame measuring approximately $4 \frac{3}{3} \mathrm{in} \times 3 \frac{3}{3} \mathrm{in} \times \frac{5}{6} \mathrm{in}$.

The normal polarizing voltage is 250 (test $1,000 \mathrm{~V}$ ) and the capacitance $0.001 \mu \mathrm{~F}$. The frequency response, according to the makers' published curve, is $7,000-15,000 \mathrm{c} / \mathrm{s}, \pm 5 \mathrm{db}$.

## Transistors

A RANGE of transistors is now available from Mullard to equipment designers who wish to carry out experimental work. These are twopoint contact types. the OCSO and


New transistors from Mullard.
OC51, and three junction types, the OC10, OC11 and OC12.

The point-contact types are readily available at a price comparable with that of mains subminiature valves. Their point spacings have been made different so that the two transistors have markedly different characteristics. The OC5l has a better high frequency characteristic than the OC50. The OC50, however, operates more satisfactorily than the OC51 in the "collector bottomed" condition.
The junction transistors, OC10, $\mathrm{OCl1}$ and OCl 2 are designed for economy in power supplies and, in

Körting electrostatic high-frequency loudspeaker.

both amplifier and oscillator circuits, they will operate from h.t. supplies as low as 1.5 volts, with current consumptions of a similarly low order. Under suitable conditions they will work with h.t. supplies of only a fraction of a volt.

The OCll is a general-purpose amplifier which, in an earthed emitter circuit, gives a current amplification factor of 17. Under similar conditions, the OCl 2 gives a current amplification factor of 30 , and is intended primarily as an output transistor, although it can also be used in amplifier circuits. The OCl0 is a low-noise version of the OC11 and is intended for use in the early stages of high-gain amplifiers.

Mullard's address is Century House, Shaftesbury Avenue, London, W.C.2.

## Nickel for Valves

TWO new grades of nickel in wrought form are now available for valve manufacturers from Henry Wiggin \& Co., Birmingham, 16. Both grades have a minimum nickel content of 99.5 per cent and are low in the volatile elements such as magnesium. Grade H.P.A., which has a low silicon content ( 0.03 per cent max.), is useful for the cathodes of valves which must have a very long life with low emission current. In Grade H.P.B. the silicon content is 0.15 to 0.25 per cent with the object of increasing activation and emission. The makers warn that care must be used with H.P.B. nickel, as under certain conditions high interface impedances may build up between the cathode sleeve and the coating material.

| Element | H.P.A. | H.P.B. |
| :---: | :---: | :---: |
|  | (per cent) | (per cent) |
| C | 0.10 max | 0.10 max |
| Cu | 0.04 max | 0.04 max |
| Fe | 0.05 max | 0.05-1.0 |
| Mn | 0.02 max | 0.10 max |
| Mg | 0.01 max | 0.01 max |
| Si | 0.03 max | 0.15-0.25 |
| Al | 0.01-0.05 | 0.02 max |
| S | 0.005 max | 0.005 max |
|  | 99.5 min | 99.5 min |

## Miniature Attenuator

THE illustration shows a miniature stud-type attenuator no larger than a two-shilling piece ( $1 \frac{1}{8} \mathrm{in}$ ) and measuring 115 in in depth behind the panel.

It is described as the Type $M$ and is intended primarily for audio applications in portable equipment and where space is at a premium. In certain circumstances it could be employed in medium- or carrierfrequency circuits up to about $4 \mathrm{Mc} / \mathrm{s}$.

Painton 1 -W high-stability carbon resistors are fitted and the attenuator can be arranged as either a 10 -step twin-arm network (bridge $T$ ) or a


Painton miniature stud attenuator, Type M.

20-step single-arm network (unbalanced potentiometer). Alternatively, it can be arranged as a fader control.

The knob and dial are so designed that only the portion of the network in circuit is indicated by the figures uncovered by the skirt of the knob.

The makers are Painton \& Co., Lid., Kingsthorpe, Northampton.

## Hearing Aid with Transistor

CONSIDERABLE saving in h.t. battery consumption has been effected in the Amplivox Model J2 hearing aid by the use of a junction transistor in the output stage. Low-consumption valves are used in the voltageamplifying stages, but the greater part of the normal h.t. current is saved by using the transistor; actually a reduction from 0.35 mA in the all-valve Model J to 0.04 mA in the J 2 , or a ratio of $120 / 1,000$ hours life. A single Mallory mercury ceil (1.3V) is used for l.t. and here the economy is useful, but less spectacular, 20 mA to 12.5 mA or 50 to 80 hours. The air-to-air gain of the J 2 is 47 db with E9L high-sensitivity earphone, compared with 53 db in the all-valve J hearing aid,

The price of the J2 is $£ 442 \mathrm{~s}$ and the makers are Amplivox, Lid., 2, Bentinck Street, London, W.L.

Amplivox Model 12 hearing aid with junction tronsistor output stoge.



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your disposal the complete service your disp


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## The NEW <br> Model T. 635 Amplifier



This outstanding 30 watt high fidelity amplifier provides all the features needed to cover the large maiority of Sound installations. It is designed for A.C. operation and can also be used on batteries with a 6 volt adaptor unit. Inputs for 2 Microphones, and one Gramophone pickup, with individual mixing controls. Separate controls for Bass and Treble boost.
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## Random radiations

By "DIALLIST"

## At Long Last!

Unless Anythinis so entirely monstrous as the shelving of the Television Advisory Committee's all-but-unanimous report should take place, f.m. is at long last recognized as the victor in the great (and far too protracted) controversy over the best modulation system for v.h.f. broadcasting. No one who had been receiving the Wrotham broadcasts, as no doubt all members of the technical sub-committee had, could have failed to realize the vast superiority of frequency modulation as a defence against motor-car-ignition interference, which is the only really important kind of interference on the v.h. frequencies. Given a receiver so arranged that a quick switch-over from one system of modulation to another can be made, two or three short periods of listening should convince most people that a.m. with limiter is a bad second to f.m., with plain a.m. very much an also-ran. Since it has been shown that a nation-wide f.m. service could be installed at little more than one-third the cost of a similar a.m. service, and that the latter would call for 44 per cent more stations, with a power output more than $3 \frac{1}{2}$ times as great and costing $2 \frac{1}{2}$ times as much to run, I can't help feeling very much in sympathy with the Committec's majority report.

## Shaggy Dog Story

Unlike Most of the others, this Shaggy Dog Story is true, cross my heart; see that wet, see that dry. Hailing me as I was passing her house the other day, my friend Mrs. X wafted me into her drawing room. "Have you ever heard," she asked, "of a dog that was radio-active or something? "

## ??????

"Well, I think Wuffles must be." Wuffles is an Old English Bobtail. "Yesterday afternoon, when the children and I were watching Muffin, he just gave himself a shake over there and the television set immediately went mad."
"What happened?"
"First, the picture shrank to halfsize; then it blacked right ous. I tried twiddling knobs and things,
but that didn't do any good. There were so many flashes and crackles that I switched off and 'phoned the serviceman. He took the set away this morning, saying that its innards looked as if it had been struck by lightning."

I asked whether by any chance Wuffles was wet when he performed that devastating shake. He was! He'd been brought in at the back door by her husband, after a walk in the rain, and had escaped and made his way into the drawing room before he could be dried off. Most of him was behind the television set when he shook himself. One could picture large drops of water being propelled through the louvres while a mist of droplets was borne in by convection currents. Wet dogs, it seems, should be kept far from TV receivers; the shaggier the farther.

## A C.R.T. Weak Spot

A Cause of television c.r. tube breakdown which seems to be growing increasingly common is the development of a "short" between heater and cathode. The fault is often intermittent-and that kind is the most exasperating of all. It can be cleared in suitable cases by the use of an isolating transformer; and
the fact that special transformers are readily available shows that these shorts are not exactly rarities. When one considers the cost, including purchase tax and labour charges, of replacing one of the large tubes so popular today, one can't help feeling that a very special effort should be made to overcome any known weaknesses in them. Here is one which calls urgently for attention. For it is clearly not always possible to use an isolating transformer; and even when it is possible there are dealers here and there who can't, or won't, undertake the job.

## But W'e Likisd it

If You Want some fun, get some friends who are old hands at wireless to ransack their junk boxes and lumber rooms and join with them in building a receiver as nearly as possible like those used to reproduce broadcasting in its carly days in this country. Then, having made the set, you may spend entertaining hours in listening to its performances and in wondering how people in those days could possibly have thought them endurable, let alone beautiful. Some friends and I did just that a year or two ago. We were able between us to bring together a fine collection of period-piece components. My own contribution was three typical triodes, which had actually had very little use. The triode was in those days the most complex valve available; and when I say typical I mean that $r_{a}$ was of the order of


40,000 and $g_{\mathrm{in}}$ round about 0.225 $\mathrm{mA} / \mathrm{V}$. One man proudly produced a loudspeaker famous in its day for the quality of its reproduction: an Amplion "Lion," with a conical metal horn, about 12 in long and 4 in in diameter at the output end, fixed to the case of a rather large telephone receiver. Others contributed an assortment of coils, transformers, condensers and resistances (to call the last two capacitors and resistors would be an anachronism) and we were all set to go.

## Loud and Clear!

The circuit, selected from a handbook of the period, consisted of a tuned-grid, tuned-anode r.f. amplifier, held down by the application of positive grid bias via a potentiometer; a leaky-grid detector, with reaction of the variably coupled coil type: and an a.f. stage, coupled to the detector by a $5: 1$ transformer and working with about 2 volts negative grid bias, which was as much as it would stand with the historically correct $90-\mathrm{V}$ h.t. But, younger readers may exclaim, what about the output stage? Dear younger readers. that a.f. stage was the output stage! The small power valve didn't come along until broadcasting had been going on for some time. The same Marconi "R," or Mullard "Ora" or Cossor "Tin-hat" valves had to serve in each and every stage of early broadcast receivers. Also included in our collection were several numbers of the popular radio magazines of the early 1920s. These confirmed our recollection that the two qualities then most esteemed in a broadeast receiver were the number of stations that it would "get" and the volume with which it would bring in the local station: "Even at 50 yards from the speaker," wrote one enthusiast in the correspondence columns, "every note is clear and free from all distortion!" Bearing in mind the conditions prevailing in that output stage-and indeed in all those stages -you may feel that this writer had been a little over-enthusiastic. Make up a set on the lines suggested and you'd wonder how anyone could have endured its output for a moment, even with the volume turned right down. Will my successor in the Wireless World of thirty years hence, I wonder, make similar experiments by reconstituting a 1954 wircless or television set? And, if he does, will he marvel in the same way at the crudities which our eyes and cars must have accepted without protest?

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## RANGE OF CONTROL-KNOBS



A NEW Range of Instrument Knobs and Dials. Manufactured in the finest-grade polished Bakelite, with frosted aluminium " Silver-Dial" dials.


| List No. | Item | Dimensions. etc. |
| :---: | :---: | :---: |
| K. 401 | Knob | $\begin{aligned} & 1 y^{3}{ }^{\prime \prime}(29.4 \mathrm{~mm}) \\ & (1 / .5 \mathrm{~mm}) \mathrm{hieh} \end{aligned}$ |
| K. 405 | Skirt |  |
| K.411 | Dial | $\begin{aligned} & 2^{m}(50.8 \text { mm. }) \times 21 \\ & \text { S.W.G., engraved } 0-10 \text { over } \\ & 270 \end{aligned}$ |
| K.411/P | Dial | dicto, not engraved |


| List No. | Item | Dimensions. erc. |
| :---: | :---: | :---: |
| K.402 | Knob | $\begin{aligned} & 18^{*}(41.3 \mathrm{~mm} .) \text { o } \times \text {. } \\ & (19.9 \mathrm{~mm} .) \text { hizh } \end{aligned}$ |
| K. 406 | Skire | $2 \frac{1}{10}{ }^{*}\left(52.4 \mathrm{~mm}\right.$.) $\otimes \times 1 \mathrm{Na}^{5 \times}$ $(5.9 \mathrm{~mm})$ thick |
| K. 412 | Dial | $2 \frac{3 .}{\frac{3}{*}}(69.9 \mathrm{~mm}$.) $\varnothing \times 21$ <br> S.W.G., engraved $0-100$ over 180 |
| K.412/P | Dial | ditco, not engraved |



Further details available in the NEW 114 page Catalogue. Price $/ /=$ pose free. Ref. $192 / \mathrm{WW}$.

| List No. | Item | Dimensions, erc. |
| :---: | :---: | :---: |
| K. 403 | Knob | $\begin{aligned} & 2 \pi(60.3 \mathrm{~mm}) \text { is } \times{ }^{2} \\ & (24.6 \mathrm{~mm} .) \mathrm{high} \end{aligned}$ |
| K. 407 | Skirt | $\begin{aligned} & 3^{\prime \prime}\left(76.2 \mathrm{~mm} .1 \text { a } \times 15^{\prime \prime}\right. \\ & (59 \mathrm{~mm} .1 \text { thick } \end{aligned}$ |
| K. 413 | Dial | $4(101.6 \mathrm{~mm}.) \not \times 21$ <br> S.W.G., engraved $0=100$ over 1? $0^{c}$ |
| K.413 P | Dial | ditto, not engraved |



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## Alice in Solidconductorland


#### Abstract

"A current of electricity from $A$ to $B$ consists of electrons moving from $B$ to $A$ and/or positive ions moving from $A$ to B. If the current is flowing in solid material, the ions are the fixed material itself, so cannot move and the current consists wholly of electrons."


The above definition of an electric current in a solid conductor, which is quoted from the article in the January issue of $W . W$. dealing with the speed of electricity is accurate beyond the shadow of a doubt. It does, however, have a seeming oddness if considered carefully and my warped mind has sometimes wondered if this is not due to the fact that in pre-electronic days an unlucky guess was made about the direction of electric current flow and we still cling stubbornly to it.
"What is going on ?" said Alice to the Red Duchess as she pointed to a rather disorderly mob of soldiers streaming out of one of the castle towers which had a big B over the gate and entering another tower labelled A. "Oh," said the Duchess, " just troop movements; I have decided to evacuate tower A and house all the troops in tower B." "But," protested Alice, " They're not obeying your orders; they are evacuating B and marching over to A." "Nonsense, child," said the Duchess following the direction of Alice's pointing finger. "What you see are merely soldiers; the troops which are moving in the opposite direction are not really moving at all."
"Then do troop movements from tower A to tower B consist of soldiers marching from $B$ to $A$ and nothing moving from A to B?" asked Alice. "Certainly" said the Duchess. "What a strange thing" commented Alice. "Wouldn't it make things easier to call the soldiers troops, and then troops and soldiers would always be marching in the same direction because they would, in fact, be one and the same thing?"
"Maybe to your simple mind it would make things easier" said the Duchess with some asperity, "but you have no respect for tradition; a century or more ago when troops were first discovered my great-grand-
father, who, being blind, mistook the directions in which they were marching and gave orders that the castle entrances were to be labelled exits and vice-versa., As everybody else was blind too" continued the Duchess, "it didn't matter much and the troops didn't care anyway as they knew the way in and the way out by the same sort of natural law or homing instinct which enables a male hippopotamus, even when he is blindfolded, to distinguish a female hippopotamus , from another male hippopotamus."
"But nobody is blind now" protested Alice. "Of course not, child," said the Duchess chidingly, "but out of respect for my great-grandfather's


"What is going on?"

memory we still like to pretend he was right even though it gets us into all sorts of difficulties, so when soldiers are moving from B to A we always pretend that troops are moving from A to B even though troops don't really move; at any rate not in Solidconductorland in which we live.
"Well" said Alice "I call it silly; we have nothing like it in our country." "Rubbish" snapped the Duchess beckoning to her executioner, "What about your summer time? What is obviously midday you pretend it is one hour past midday although everybody knows full well that it really isn't." "Off with her head," she shouted to the executioner.

## A Unique Opportunity

Nobody seems to have pointed out what a magnificent opportunity the proposed introduction of competitive television woild offer for
starting the higher definition of which we have heard so much from those who criticize the B.B.C.'s 405line system. Now at last a unique opportunity arises for confounding these critics or justifying their views.

Before any competitive service could start we would all, in any case, have to buy adaptors or new sets if we want to receive the programmes. Those who could not or would not go to the expense of this would be no worse off than they are at present, for they would still be able to receive the B.B.C. programmes. But if people were willing to go to the expense of buying an adaptor or a new set to receive the alternative programmes, then why not give them full value for their money in the shape of a picture of the highest definition which modern technique can provide?

Naturally this would be very hard lines on the B.B.C.-about 625 of them, I suppose-and the Corporation would be torn between its desire to radiate a picture of equally high definition and the necessity of keeping faith with owners of existing sets who didn't want to spend money on new receiving gear. But I feel that owners of existing sets would be so jealous of the technically better picture of "competitive" viewers that they would soon alter their views and their viewing apparatus, thus freeing the B.B.C. from its obligations and enabling it to compete on equal technical terms with, or even to go one better than, its competitor by radiating stereoscopic TV, to say nothing of colour.

## No Suppressors Needed

At this time of the year the big toy shops in all our large towns usually make an intensive drive to sell model electrically-operated railways, which they show in operation. These are prolific sources of interference to neighbouring TV sets as, so far as I can see, no attempt is made to fit suppressors.

It may be argued, of course, that these trains are not being demonstrated in normal television hours, but one cannot expect the average modern child to discontinue the use of his model railway when TV starts. At least, such was mv opinion until recently, when I was discussing the problem with a specimen of the 1963 National Service class. To my surprise he pointed out with withering contempt that he and others of his age group were the backbone of the B.B.C.'s TV audience and would not foul their own nest. I felt completely silenced and suppressed.


An inexpensive yet precision instrument designed especially to meet the exacting needs of the modern service With six and laboratory technicing covering $50 \mathrm{Kc} / \mathrm{s}$. to frequency ranges coracy is better than $80 \mathrm{Mc} / \mathrm{s}$. , its acale reading.

$\begin{array}{ll} \pm 1 \% & \text { of the scale readig. } \\ 50 \mathrm{Kc} / \mathrm{s} .-150 \mathrm{Kc} / \mathrm{s} . & 1.5 \mathrm{Mc} / \mathrm{s} .-5.5 \mathrm{Mc} / \mathrm{s} .\end{array}$ |  | $50 \mathrm{Kc} / \mathrm{s} .-150 \mathrm{Kc} / \mathrm{s}$. |
| :--- | :--- |
| $150 \mathrm{Kc} / \mathrm{s} .-500 \mathrm{Kc} / \mathrm{s}$. | $5.5 \mathrm{Mc} / \mathrm{s} .-20 \mathrm{Mc} / \mathrm{s}$. | $500 \mathrm{Kc} / \mathrm{s} .-1.5 \mathrm{Mc} / \mathrm{s}$. $20 \mathrm{Mc} / \mathrm{s} .-80 \mathrm{Mc} / \mathrm{s}$. Scale sub-divisions provide more than adequate discrimination for use in television circuits. Note the starred features below, which combine than maintain a min $20 \mathrm{Mc} / \mathrm{s}$. and less than $3 \mu \mathrm{~V}$ between $20 \mathrm{Mc} / \mathrm{s}$. and $80 \mathrm{Mc} / \mathrm{s}$.

## Come <br> Aqain

## Radio-Electronic Men!

Just as you have been coming since 1945 to the IRE National Convention and Radio Engineering Show - coming by the thousands, 35,642 in ' 53 - so come again to see and hear all that is new in the engineering advances of your industry.

## A Fifty-four in '54!

- 243 scientific and engineering papers will be presented, skillfully grouped by related interests into 54 technical sessions. More than half these sessions are organized by IRE Professional Groups, thus making the IRE National a federation of 21 conferences in one. The whole provides a practical summary of radio-electronic progress.

A 600 Exhibitors "spotlight the new!" - A mile and a half of exhibits line the avenues of this show, intriguingly named for the elements of radio - such as "Instruments," "Components," "Airborne," "Radar," "Transistor," "Audio," "Microwave," etc., filling the four acres of the great Kingsbridge Armory to capacity. An expanding radio industry shows why it is growing by proving how engineering research pays out in new products. The exhibits themselves are an education, condensed to one place - reviewed in four days.


Waldorf-Asterite and Kingsbridge Armory

Only the combined facllities of the Waldorf-Astoria Hotel, plus the three great halls in the Kingsbridge Armory, seating 906,720 , and 500 respectively, are able to keep pace with the increased technical papers program of the IRE Convention. The show had to move because the U. S. Treasury took over Grand Central Palace. The immense Kingsbridge Armory, connected to the very satisfactory Lexington Avenue Hotel area by direct express subway, serves well to expand the already outgrown exhibit facilities of the Palace and pro-
vide space for 200 new firms to exhibit, as yell as seat greater audiences at the high-interest sessions. In addition to the subways, free busses leave the Waldorf every ten minutes in which you may travel in the congenial company of fellow en gineers, direct to Kingsbridge.

A Admission by registration onlyl Registration serves for the four day period. It is \$1. for IRE members, \$3. for non-members, covering sesslons and exhibits. Social events priced separately
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and
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New production methods and exhaustive tests assure users of Mullard Special Quality Valves of the definite standard of reliability laid down in Ministry of Supply (Air) Sperifications.
Large-scale production of valves which are capable of withstanding such arduous conditions of shock and vibration presents obvious difficulties. These problems have been vigorously investigated over the past five years by Mullard who have now evolved techniques which make Special Quality Valves an economic reality for a wide variety of electronic equipment.
The first types in a comprehensive Mullard range are now generally available and are listed below. Details of the availability of types at present under advanced development will be announced soon.

| Mullard Special Quality Type No. | Description | Electrical Equivalent |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  | Mullard | American | British Services |
| M8079 | Double Diode with separate cathodes | E891 | 6AL5 | CV140 |
| M8081 | Double Triode for use as R.F. amplifier or oscillator | ECC91 | 6 J 6 | CV858 |
| M8082 | Output Pentode | EL91 | - | CV136 |
| M8083 | High-slope R.F. Pentode | EF91 | - | CVI38 |
| M8100 | Low-noise high-slope R.F. Pentode | EF95 | 6AK5 | CV850 |

## $\underset{\text { SPECIAL }}{\text { TESTED TO }} \underset{\substack{\text { SPECIFIED } \\ \text { EXTREMES }}}{\substack{\text { VULITITY }}}$


I. Dubilier High Stability Resistors have a low temperature coefficient of resistance not exceeding minus $0.03 \%$ for values up to $100 \mathrm{~K} \Omega$ and not exceeding minus $0.04 \%$ bye per ${ }^{\circ} \mathrm{C}$ for values above 100 K to $1 \mathrm{~m} \Omega$.
${ }^{3}$ 2. The voltage coefficient is low and does not exceed $0.002 \%$ per volt applied. The noise level is very low and is less thian $0.5 \mu \mathrm{~V}$ per volt applied and in most instances is less than $75 \%$ of this value.
3. The resistors have-excellent high frequency characteristics with low capa-

- citance of the order of 0.5 pF and with an excellent inductance to resistance ratio.

4. The long term stability is excellent in both working and idling conditions and is of the order of $1 \%$. The resistors are of particularly robust construction and suitable for use under the most stringent conditions.

Catalogue HSR 753 giving details of the full range of Dubilier


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 TAPE RECORDING EQUIPMENT

Model 2A provides at a reasonable cost an instrument approaching professional standards - its specification commending it especially to those engaged in educational and cultural pursuits.


Equipment Type YDC is a simultaneous dual-channel Recorder/Reproducer offering special facilities to those engaged in analysis problems in the medical, aeronautical and scientific fields.


Ferrograph Model YD. An instrument version of the famous $2 \dot{A}$ having balanced 600 ohm inputs and outputs, a tropicalised amplifier, and a performance up to professional standards. It can be rack mounted.


A switching unit developed for use with recorders type YD and YDC. It switches on the recorder only when a signal is present and enables the traffic of many hours to be recorded on a normal length reel.

Ferrotape is now freely available in the following sizes: -

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\begin{array}{cccc}
200 \mathrm{ft}, & 600 \mathrm{ft} . & 1,200 \mathrm{ft} . & 1,750 \mathrm{ft} . \\
12,6 & \epsilon 1-6.9 & \epsilon 2.5 .0 & \epsilon 3.3 .0
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Once again, Finland's traditional foresight has carried her people forward in the march of progress.

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The Radio Corporation of America is proud of the opportunity to have cooperated in this significant achievement and joins Havulinna Oy, its distributor in Finland, in saluting the Government of Finland, and the Finnish Broadcasting Company . . . a pioneer in FM networking on the European continent.

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humidity and corrosive agents contribute further to the success of this new epoxy resin for potting or casting applications. 'Araldite' complies with the requirements stipulated for the sealing of Service equipment. Our illustration of a transformer potted in 'Araldite' is published by courtesy of the makers, Messrs. Evershed \& Vignoles Ltd., who also use the same resin for sealing resistances and valve assemblies.
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Designed for the measurement of any frequency in the range $10 \mathrm{c} / \mathrm{s}$ to $1 \mathrm{Mc} / \mathrm{s}$ with a basic accuracy of $\pm 0.005 \% \pm 0.1,1.0$, or $10 \mathrm{c} / \mathrm{s}$.

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It is light, compact and simple to operate and can be supplied either in a transportable case or for standard $19^{*}$ rack mounting.

## Specification

FREQUENCY COVERAGE: $\begin{aligned} & 50-20.000 \mathrm{c} / \mathrm{s} \text { in two } \\ & \text { ranges }\end{aligned}$
FREQUENCY STABILITY: Better than $\pm 1 \%$
INPUT LEVEL:

FREQUENCY
Between -20 and +20 db relative to 1 mV into 600 ohms adjusted by a 5 step attenuator and fine control of $0-10 d b$
Flat to within $\pm 1$ db over the whole working range.
LEVEL STABILITY: $\pm 1$ dbifmains supply voltage remains within $\pm 5 \%$ ACCURACY OF Harmonics below fundamental
HARMONIC
MEASUREMENTS :
to $-55 \mathrm{db} \pm 5 \%$
to $-75 \mathrm{db} \pm 10 \%$
INPUT
IMPEDANCE:
SELECTIVITY:

HUM LEVEL: POWER SUPPLY: POWER
CONSUMPTION :
DIMENSIONS:
WEIGHT:
or $\pm 1 \mathrm{db}$ whichever is the greater.
$100,000 \mathrm{ohm}$ unbalanced (greater than 25,000 ohm, balanced, can be supplied to order)
Constant over the frequency range. 2 stage crystal filter, mid band frequency $100 \mathrm{Kr} / \mathrm{s}$. Flat top ( $\pm 1$ db) pass band extending over $10 \mathrm{c} / \mathrm{s} \pm 1 \mathrm{c} / \mathrm{s}$. $-55 d b$
$110 / 115$ and 200/250 volts $40 / 60 \mathrm{c} / \mathrm{s}$
Approx. 60 watts
In case $20^{\circ} \times 9^{\prime \prime} \times 8 \frac{1}{2}^{\prime \prime}$
31 lbs. approx.

## ?

THE GFT. 560 is a.IkW channelised transmitter with a frequency range of $1.5-30 \mathrm{Mc} / \mathrm{s}$. It consists of three basic cabinets-r.f. unit, modulator unit, and power supply unit-combinations of which can be used to provide multi-frequency working as well as a number of different types of emission. The wave change facilities of the transmitter are both rapid and reliable-a valuable asset when the operating frequency is changed many times each day.
The GFT. 560 is fully tropicalised, and its unit construction facilitates future expansion of the initial installation, should the need arise.
For use in conjunction with the GFT. 560 there are ancillary units that enable the transmitter to be remotely controlled over a two wire telephone circuit : operational adjustments are dialled to the transmitter. The versatility and reliability of this new Mullard transmitter make it particularly suitable for h.f. en-route, ground-to-air services and point-to-point communication networks. A team of Mullard communication engineers is available to advise on the use of the GFT. 560 in such applications. They will also assist in planning complete communication systems, if required.

ABRIDGED DATA

## Frequency Range 1.5 -

 $30 \mathrm{Me} / \mathrm{s}$Frequency Stability To Atlantic City, 1947, standards Power Output / kW Types of Emission c.w., mac.w., telephony, frequency shift, single and independent sideband. (AI, A2, A3, F1, A3a and A3b) Output impedance 600 ohms balanced twin feeder Power Supply $400 \mathrm{~V}, 50$ $60 \mathrm{c} / \mathrm{s}$, 3-phase

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possible in the design and production of R. \& A. Reproducers to ensure that initial performance is maintained over long periods of use. A typical example is the patented method of moulding phosphor-bronze strip voice-coil leads between the two layers of Bakelised linen which form the centring member.
We don't suggest that this works miracles, but it does avoid the buzzes and rattles commonly caused by flexible leads anchored to the cone; and it prevents both distortion of the cone when eyelets are inserted, and the effects of localised heat during the subsequent soldering operation.

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No. 2 " SYMPHONY " AMPLIFIER as No. I but with IO-watt Push-Pull triode output and triodes throughout. Woden mains and output transformers and choke. Full provision and power for Tuner. Output tapped 3, 7.5 and 15 ohms. Competes with the most expensive amplifiers on the market yet costs only 15 gns . (carriage $5 /$-). trers on the market yet costs only 15 gns. (car
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-SYMPHONY" AMPLIFIERS with REMOTE CONTROL. Both the above model Amplifiers are available with all controls on a separate Control panel with up to 4 feet flexible cable which simply plugs into the amplirier. Enables the Amplifier oroper to be sat in the bottom of a cabinet whilse the conerols are mounted conveniently higher up. Extra cose 2 gns.

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your problem with our Chief Engineer (avallyour problem with our Chief Engineer (avali6 p.m.). Or , write enclosing $2 \nmid \mathrm{~d}$. stamp. This Technical Guidance Service will cost you nothing and save you pounds.
NOTE: Regarding Pick-up heads to take standard or miniature thorns for 78 r.p.m., these can be supplied with any of the single-record Gram Units or Auto-changers sold by us. if desired
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by E. D. Daniel, M.A., A.M.I.E.E. anai
P. E. Axon, O.B.E., M.Sc., Ph.D., A.M.I.E.E. Research Department, B.B.C. Engineering Division The recording characteristic of a magnetic tape system can be defined in terms of the magnetic induction at the tape surface. A method of measuring surface induction is described which employs a non-magnetic conducting loop as a reproducing head. Using tapes calibrated by means of this device, the action of conventional magnetic-core reproducing heads is examined and various empirical expressions are deduced for their response. It is shown that if suitable corrections are applied conventional heads having either short or very long gaps may be also used for standardisation. Finally, the effects of various common imperfections and maladjustments in all types of reproducing heads are examined in detail.

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Head of Engineering Training. Department, B.B.C. Engineering Division

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| Typical D.C collector voltage (V) | 2 | 2 | 2 |
| Typical collector current (mA) | -0.5 | -0.5 | -2 |
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## The presentation

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20 amps.

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HS2. $250-0.250$ v. $80 \mathrm{~m} / \mathrm{a}$.
HS3. $350-0-350$ v. $80 \mathrm{~m} / \mathrm{a} ., 19 / \mathrm{m}$. H30. $300-0-300 \mathrm{v} .80 \mathrm{~m} / \mathrm{a}$
H\$2X. 250-0-250 v. $100 \mathrm{~m} / \mathrm{a} ., 21 / \mathrm{m}$. H\$75. 275-0-275 v. 100
HS30x. $300-0-300$ v. $100 \mathrm{~m} / \mathrm{a} ., 21 / \omega_{\text {. }}$ HS3x. $350-0$ - 350 v

## Fully Shrouded

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F $\$ 30$. $300-0-300$ v. $80 \mathrm{~m} / \mathrm{a}$, , 21/.. FS 3 . $350-0.350$ v. $80 \mathrm{~m} / \mathrm{a}$..
FS2X. $250-0.250$ v. $100 \mathrm{~m} / \mathrm{a}$, , 23/. FS75. 275-0.275 v. 100

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FS50. Output $450-0-450$ v. $250 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v} .2$ amps., C. $\mathrm{T}, 6.3$ v 4 amps., C.T. 5 v, 3 amps. Fully shrouded
F35X. Outpur $350-0-350$ v. $250 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v} .6 \mathrm{amps} ., 4 \mathrm{v} .8 \mathrm{amps}$ 4 v. 3 amps., 0-2-6.3 v. 2 amps. Fully shrouded
FSI60X. Oueput $350-0-350$ v. $160 \mathrm{~m} / \mathrm{a} .6 .3 \mathrm{v} .6 \mathrm{amps} .6 .3 \mathrm{v}$
3 amps., 5 v. 3 amps. Fully shrouded
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6 amps 5 y 3 amps. Fully shrouded

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The circuit Is the latest type TRF using 3 Valves and Metal Rectifiers for operation on 200/250 volt A.C. mains. Waveband coverage is $180 / 550$ metres on medium wave and $800 / 2,000$ metres on long wave. The Dial is illuminated and the Valve line up is : 6K7-H.F. Pentode, 6J7-Detector and 6V6-Output. The attractive Cabinets to house the Receiver, size 12 in . long. $6 \underset{\mathrm{l}}{\mathrm{in}}$. high, $5 \$ \mathrm{in}$. deep, can be supplied in either walnut or ivory Bakelite or wood.

1 Aerial Coil (Green Spot) with Fixing Bar ( Anode Coil Spot ) with Fixing Bar
1 Wavechange Switch
1 2-gang Variable Condenser with Trimmers

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SPECIAI PRICE F TION $12 / 6$ plus packing \& postage $1 / 6$. SECTION 3
1 Choke
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 Including ALL parts, Valves, Portable Cabinet, Bin. Loudspeaker, TapeTable, Reel of 'Scotch Boy' Tape and Rewind Spool, and Microphone.Brief Apecification: VALVE LINE-UP: RP37A First Btage, 6817 second Stage and Tone Control; GV6 Output 6X5 Rectifler; VT501 Blas and Erase Oncilator; 7193 Record Level Amplifter; GUS Magic Eye Record Level Indicator, OUNMme; Record/Playback Switch ; Treble Boost; BassBoost-on/off.
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FAST REWIND. Provision for fast rewind and forward run 18 legs than 1 min. In either direction. WRND AND REWEA MOTORS obviating friction drive.

HGR FIDELITY RECORD PLAYBACK (1 HOUR APPROX. PLAYING) The Tablefs fitted with high fidellty record playback head of new desikn wound to allowing approx. 1 hr , playing fromstanderd $1,200 \mathrm{ft}$. Reel of Tape.
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AMPLIFLER KIT (including 8in. Speaker) ...... \&11 0 0 plas 5/a pkg.farr. AMPLIFIER (already built,wired snd tested).... \&14 150 Dlus $7 / 6$ phg./cart. LANE TAPE TABLE \& REWIND SPOOL ...... 81710 0 plas $7 / 6$ pleg./cart PORTABLE CABINET (rozine covered) …... £4 19 6 plus $5 /-$ pkr./carr. MIUROPHONE
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This is credited if a complete kit of the Tape Recorder is ordered.


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Three years ago we gave you the 6 in ., 9 il .. and 12 in . Televisors which achieved zremendous popularity. Now after a considerable period of research our Technical Staff have designed a very worthy successor to these original Models.

## Brief Technical Details are as follows:

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VISION CIRCUIT. Common RF Amplifier, single valve frequency changer, two IF stages, Video Detector and Noise Limiter followed by special type of Video. Output Valve. ALL COILS PRE-TUNED ASSURING ACCURATE ALIGNMENT AND EXCELLENT BAND. WIDTH.
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The Televisor may be constructed in 5 easy stages: (1) Vision, (2) Time Base, (3) Sound, (4) Power Pack, (5) Final Assembly. Each stage is fully covered in the Instruction Book, which includes layout, circuit diagrams and point-topoint wiring instructiors.


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A handsome Walnut Cabinet that will be a fitting housing for a first-class Televisor.
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## M. \& L. WAVEBANDS

 Falve line-up: 12J7, S5LB, 1487, 35 Z 4 . Entirely transportable and tnuaualis sensiture owing to special feed-back elrcuit em-ployed. Horased in attractive ployed. Howsed in attractive
plasucablret. plastlecablnet. Cholce of 2
and Crearn. Curring handle inoorporated 200/250 A.C.ID.O. mains. Fully covered by Marufacturer's Guarantee

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 Range 30 to $40 \mathrm{Mc} / \mathrm{s}$. Contains six now Valves
$3.9 \mathrm{D} 2,1-8 \mathrm{D} 2$, 1.15 D 2 $3.9 \mathrm{D} 2,1-8 \mathrm{D} 2$,
(froquency
1.4 D 1.16 D 2
changer), (froquency
mers, 64 ceramic trim.
mamic mers, 6 coramic valve-
holders, 6 talve screenlng cans, 30 rasistors,
1-W/W Pot, Meter Mica 1-W/W Pot, Meter Mica
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Soljed $17 / 6$ plus $3 / 6$ pontage and packing.

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 SUGHTLYSOILED Yo orlginal casescomplete with 20 comptete Frequency range $18.5 \mathrm{Mole} .-$
$75 \mathrm{Kc} / \mathrm{B}$. in 5 wave. bands. \&\%/18/6. Plue 10/6 packlog and carrlage.

## POWER SUPPLY for sbove. Incorporating out-

 Lor sbove. incorporaking out-put staye. Supplles an output stace. supples an out. Which is ample for the Rills with the outputshage.
Jones pluga for connechng

Jones pluge for connecting the Power Pack to the Heceiver are Included. The ov/8 output stage combultinto the unit. Prico $85 / 5 / \%$, plus $5 /$ - packing and carriage.

We have a few Erand new Rilf5 Receivers in original cases, complete with 10 valves. Frequency range $18.5 \mathrm{Mc} / \mathrm{s}-75 \mathrm{Kc} / \mathrm{s}$ in 5 wavebonds. PRICE $\mathbb{\text { ClI/I9/6. Plus } 1 0 / 6 \text { pkg. and carr. }}$

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Prequancies covered $30.30 \mathrm{Mc} / \mathrm{s}$ ( 10.15 metres). Switched tuining, 5 prootaned apot ireq. $3 /$ VB65 (SP61). $12 / 6$.

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Frequoncles covered $40-50 \mathrm{Mc} / \mathrm{s}$ (6-7.5 metres), switched tualing. 5 Prosect positions complote with 3 VB65's 17/6.

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The ideal ahort.wave converter for T.V., variable tualing, vontains 2-EF54. 1-VR137, 37/6.

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8/6 0/m.
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T.V. PRE-AMPLIFIER

Amplifier Unit Type 208A uslug 2-VR92 val weo ruitable Amplifier Unit Type 208A uning 2-V for oppation on london
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large stock
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 LIMITED QUANTITY II32A RECEIVER UNITS COMPLETE WITH CIRCUIT 11-valve Superhet
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Superhet 100 to $124 \mathrm{Me} / \mathrm{s}$,
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drtve, R.F. and L.F. Gain. Control, etc. Circuit R.F. amg. frequenoy ohenget, oscillator and stab., 3-IF. amps., B.F.O. Det, firat andio and output
Brand new, with circuit diagram.

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POWER PACE for above completely wired and tested, whlift on Receiver chassis. Price $50 /$-plus 2/6 pkg. and cart.

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\&\$129/6. Plus $\mathrm{y} /$ - pkg., carr.,Ins. VCR517C
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A 12in. TRUVOX P.M. SPEAKER

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 VIbRATOR PACK9 v. imput 180 จ. 35 mA. outpot com pleto with ralve reoukar and leads. 39/6. Plua $5 /$ - pkg, catr.

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WILLIAMSON AMPLIFIER KIT 15 gns .
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$350-0.350$
$350-0-350$, אO mA.. 6.3
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s
, (1) 4 a. 5

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AUTO TRANSFORMERS 50 WATTS Input/Output 0-110-210-220-230-240-250
volts. Pius $1 /-\mathrm{P}$. \& P

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Designed to meet the demand for an efficient variable ratio Output Transformer. 11 ratios from $13: 1$ to $80: 1$ all centre tapped and can be used to match any output valves olther aingle- or push-pull Class " ${ }^{\text {an }}$ " " AB1" Ambination thereot. Primary Inductance 60 henries 15 watts audio 100 mA . Price $45 / \mathrm{/}$.

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Input 230 v. A.C. Output 12 v. at 1 amp. Completely input 230

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2002 Charges 6 volt accumulator at 1 mpp .
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An ontirely inaulated aryatal microphone whilch can be uniely used on A.C.ID.C. ampliflers. High impedinot. No background molse, really naiural tone. The idea

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TYPE GPX for Stondard and Long Playing
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Rexine covered, including wooden motor board. Outside diviens.; Hgt. (when closed) jilin., length 151 n. , depth inder motor board when closed 2 iln.

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The Motor, Tone arm, and Magnetic Pick-up la in one Unit, with Autorastic stop and start.
For use on $2000 / 250$ v. A.C. mains 50 cycles. Limited apuantity only. $£ 3 / 18 / 6$, plus $2 / 6$ packing and carriage.

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5 hin. acreen. 4 volt Feater. This Electrostatic Tube 15 recommended as eminently suitsble for Tolevison.

> SUPER QUALITY TELEVISION MAGNIFYING LENS
> sin. lens suitable for 6in.

18/6
sin. $1 e n s$ su
6 in. lens.
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ALUMINIUM CHASSIS 18 s.w.g. Substantisily mado Irom Bright Aluminium, with lour sides
 $10 \times 8 \times 2 / 1 \mathrm{n} . \cdots \cdots, \quad 5 / 6 \quad 14 \times 10 \times 3 \mathrm{in} . \quad . \quad 7 / 11$ $\begin{array}{cccc}12 \times 9 \times 2 \text { in. } & \ldots . . & 7 /- & 16 \times 10 \times 3 \mathrm{in} . \\ 14 \times 9 \times 2 i \mathrm{in} . & \cdots .6 & 8 / 3 \\ 16 \times 8 \times 21 \mathrm{in} . & 8 /=\end{array}$

ALUMINIUM PANELS $18 \mathrm{s.w} \cdot \mathrm{~g}$.


$1 / 3$ $1 / 8$ $2 / 2$ $2 / 8$ $3 / 2$ $3 / 8$ 418 $5 / 2$

$7 \times 4 \mathrm{in}$.
$91 \times 4 \mathrm{in}$.
$10 \times 7 \mathrm{in}$.
$12 \times 7 \mathrm{in}$.
$14 \times 7 \mathrm{in}$,
$16 \times 7 \mathrm{in}$.
$20 \times 7 \mathrm{in}$.
$22 \times 7 \mathrm{in}$.

## $\frac{1 / 5}{1 / 5}$ $\frac{1}{1 / 5}$

H.T. ELIMINATOR AND TRICKLE CHARGER KIT
All parts to construct an eliminator to glve an output of 120 volts at 20 mA , and 2 volts to charge an accumu. lator. Uses metal rectifier, $37 / 6$.

ANTI-INTERFERENCE AERIALS
offered at a fraction of eriginal cost


The aertal ts designed for reception of long, medium and short waves, with any ordinary or communications recelver, having an input impedance greater than 1,000 The installation discriminates against locally generated electrical Interference, especially on the ghort wave bands. The equipment enables the installation of an $8.3 \mathrm{Mc} / \mathrm{s}$ flatly-tuned dipole which operates as a "T aerial on medium and long waves. The aerial and receiver transformers are intended to be interconnected with a 70 ohms co-arial cable

## COMPONENT PARTS

Aluminium Aerial Transformer Assembly. Comprising one each: Aluminium transformer, Transformer clip Rubber sucker, lin. $\times \frac{1}{\text { g }}$ in. brass screw, $4 \mathrm{AB} \times$ in. brass bolt, 4BA nut
Receiver Translormer. Complete with Insulators, cltps, etc.; Porcelain Insulators, 2 each, 60 ft . Insulated Aerial Wire, 60ft. Screened Co-Axial Dowa Lead.
LESS CO-AXIAL CABLE \& AERIAL WIRE, 15/-, plus
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COMPLETE $35 /-$, plus $1 / 6 \mathrm{pkg}$. and carr.
GARRARD R1m Drive 78 r.p.m., complete 55.19 .6
with magnetic pick-up and turntable....u $2 / 6$
Packing and carriago on the above unic $2 / 6$

## MAINS NOISE ELIMINATOR KIT

Two specially designed chokes with three smoothin condensers with circuit diagrams. Cuts out aill mains noise. Can be assembled inside existing receiver, 5/6 complete.
Germaniam Crystal Diodes. G.E.C. wire ended, 2/6.
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MARCH 1954

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# VALVES, TUBES \& CIRCUITS 

## 15. VALVES FOR V.H.F. TELEVISION RECEPTION

The advent of television transmissions in the frequency range of 174 to $216 \mathrm{Mc} / \mathrm{s}$ (Band III) will mean that receivers must be capable of receiving signals radiated at these frequencies in addition to those already transmitted in the range 41 to $68 \mathrm{Mc} / \mathrm{s}$ (Band I).
Reception of signals at the higher frequencies involves changes in the input and frequency changer stages of present-day superheterodyne receiver designs. To meet these requirements the Mullard "World Series" of television valves has been augmented by the introduction of two new types:-

PCC84-Double triode for use as a cascode amplifier in V.H.F. input stages.
PCF80-Triode pentode for use in frequency-changer stages.
As with most of the other "World Series" valves, these new types use the B9A (noval) miniature nine-pin base, have heaters suitable for inclusion in a 300 mA series chain, and can operate with a receiver h.t. supply of 180 volts.

PCC84. The double triode is designed for use as a cascode amplifier, i.e. one triode section operates as a neutralised grounded-cathode amplifier and the other section as a grounded-grid amplifier. The two sections are then connected in series across the h.t. supply, the anode of the first section having a direct connection to the cathode of the second section. The cascode arrangement results in a low noise level for the input stage being achieved in the first section, combined with high gain resulting from use of the grounded-grid connection in the second section.

PCF80. The triode pentode, having separate electrode systems for triode and pentode sections, is used in the frequency changer stage. The triode section, connected as a local oscillator, is in a Colpitts circuit. The oscillator voltage and the input signal from the PCC84 are mixed either indnctively or capacitively on the control grid of tne pentode section. This section functions as the mixer stage, its performance depending considerably upon the choice of circuit components to give optimum values of conversion conductance and input damping.

The accompanying functional diagram illustrates the basic mode of operation of these valves when used in the " front-end " circuit of a television receiver for use in Bands I and III.


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## Measurement of the Characteristics of Television Aerials

It is necessary to measure the performance of television aerials to enable the relative merits of different installations to be compared. The measurements must be related to the performance of an average aerial installation and it is well known that the presence of objects near to an aerial will change its characteristics. It is impossible, however, to define an average house on which the measurements should be carried out, and it would be extremely laborious to measure the characteristics on a large number of houses, and average the results. It is for this reason that all measurements of the characteristics of aerials should be carried out above flat ground, well away from surrounding objects. It must be assumed that in a practical installation there will be an equal chance of surrounding objects enhancing or reducing the performance of the aerial. Measurements on an open site will then simulate the conditions on an average house.

It is possible to make a direct measurement of the gain of an aerial, i.e., to measure the amplitude of the signal delivered by the aerial to the receiver, and in particular to make a comparison between any aerial and a half-wave dipole. A half-wave dipole is chosen as a reference for gain measurements for several reasons. It is simple to construct and the relation between the strength of the incident signal and the voltage across its terminals can be determined from theoretical considerations. It is ideal for horizontally polarized signals, but in the case of vertically polarized waves, as in the majority of television transmissions in this country, there is a complication from the influence of the mast supporting the dipole. This can be overcome, however, by using either a "sleeve" dipole in which the supporting mast is inside the lower element, or a unipole above a horizontal reflecting plane. In practice misleading results are sometimes obtained when aerials are compared with a normal dipole, due to the influence of the mast. So as to overcome this without resorting to a special dipole, measurements are often made relative to " H " type aerial.

The gain measurement of an aerial can be carried out as follows. An open site on flat ground is chosen well away from trees, houses, telephone wires and other
objects which might cause. reflections of the signal. An ideal site is a disused airfield, but unfortunately this is often not available in a given area. The dipole aerial is now erected to a height of about 3 oft. and moved horizontally over a short distance to determine if there is a standing wave pattern present. The aerial under test is now raised in place of the dipole and the direct comparison made of the signal.


Junior Multirod Band 1 3-Element Array.

Two methods are available for measuring the signal amplitude. In the first, either the mean or the peak white leve of the television signal is measured, the measurements always being carried out when Test Card "C" or some steady picture is being transmitted. In the second method the amplitude of the black level is measured, and observations can be made at any time during transmissions. In the latter method the black level is determined by displaying the half line pulses occurring during the frame synchronising signal on an oscilloscope. With this it is often possible to separate the main signal from any ghost that may be present.

The gain measurements described above include not only the intrinsic gain of the aerial but also any mismatch loss between the aerial and the feeder and losses in the feeder itself. Error due to losses in the feeder can be eliminated by using the same length of cable for the dipole and for the aerial being tested. Mismatch losses can only
be evaluated from measurements of the impedance of the aerial. It is possible, however, to deduce the intrinsic gain of an aerial compared with that of a dipole from measurements of its directional characteristics. Experimental measurements of aerial gain by the two methods are in close agreement.

Another important measurement that is required on an aerial is its directional characteristics or its ability to reject signals arriving at the aerial in a direction different from that of the wanted signal.
Here again the measurements must be carried out on an open site, and it is essential that there is no standing wave pattern. In fact the most sensitive method of detecting a standing wave field is to measure the apparent polar diagram of an aerial which is known to have a very deep minima. If a standing wave field is present then the measured polar diagram wili be distorted and the depth of the minima will be reduced. The supporting pole should always be included when aerial measurements are made, as removal of this will change many of the aerial's characteristics.
Two other factors must also be determined in order to anticipate the overall performance of an aerial. These are its impedance and band width. The various methods available for the measurement of impedance will not be discussed here. It can be pointed out, however, that the matching requirements for a receiving aerial are far less stringent than those required for a transmitting system. As regards band width, an estimation of the ability of an aerial to accept the required range of frequencies can be ascertained both from impedance measurements and from visual observations on Test Card "C."

Measurement of the above properties of aerials give a very useful yardstick for the comparison of aerials. They are, moreover, usually in agreement with comparisons made over a large number of practical aerial installations.

## Band I Aerials

In this and future announcements, we will refer to the T.V. aerials in our present range, which are designed for the reception of existing B.B.C. transmissions, as band I aerials.
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# vortexion <br> TAPE RECORDER 



THE FOLLOWING TECHNICAL REPORT
by

# P. WILSON, M.A. 

APPEARED IN THE JANUARY ISSUE OF

"THE GRAMOPHONE"

Vortexion Tape Recorder, Type 2A Vortexion Ltd., 257, The Broadway, Wimbledon, S.W.19. (Price £84)

## Specification

Twin Track Recording Speed: $7 \frac{1}{2} \mathrm{in}$. and $3 \frac{3}{i} \mathrm{in}$. per sec.
Spools: Normally $1,200 \mathrm{ft}$., but can take up to 1,750ft.
Recording Time: $1,200 \mathrm{ft}$. spools- $\frac{1}{2}$ hr. per track at $7 \frac{1}{2} \mathrm{in}$., 1 hr . per track at $3 \frac{3}{2} \mathrm{in}$. $1,750 \mathrm{ft}$. spools$\frac{3}{4} \mathrm{hr}$. per track at $7 \frac{1}{2} \mathrm{in}$., $1 \frac{1}{2}$ hrs. at $3 \frac{3}{2} \mathrm{in}$.
Frequency Range from 50 to 10,000 c.p.s.: Amplifier response is within $1 \frac{1}{2} \mathrm{db}$.
Input: (1) Microphone (preferably Ribbon) 15 to 30 ohms with full loading on 8 microvolts. Balanced or not, so that 100 ft . unscreened cable can be used if desired. Model for Crystal microphone with input of 1 millivolt can be supplied. (2) For Radio feeder unit or Gramophone p/u or even crystal set. Fully loaded by 17 millivolts on $\frac{1}{2}$ megohm. Unbalanced.
Output: $3 \frac{1}{2}$ watts at 15 ohms, i.e. up to 7 volts. Internal speaker. Jack for external speaker or to feed any other amplifier of any impedance greater than 15 ohms.
Monitoring : Internal loudspeaker may be used to monitor during recording and there is also a meter to monitor recording level. This meter may also be used for setting bias level to suit different makes of tape. Best settings for different makes are specified.
Cabinet: Flat rexine case with carrying handle and detachable lid. All controls on tape deck or extension. Input and output comnections inset in side of case.
Dimensions: 8 fin. by $22 \frac{1}{2} \mathrm{in}$. by 15 ? t in.
Weight: 48 lb .

This instrument, though on the heavy side as portable tape recorders go, is so nicely balanced as to be fairly easily carried. It is certainly very convenient.
I collected a sample myself from the factory, took it home and started a test straightaway. It gave no trouble at all during the whole fortnight I had it in operation. During the period, I recorded test frequencies, radio and record performances and also made direct recordings of speech and music through a Reslo Ribbon Microphone.
"I paid "particular attention to detect "wow" or "flutter." since these were not mentioned in the makers' specification. I was aware that the Wearite Tape Deck which is used in this model has a good reputation in this respect, but my sample was even better than the standard: by ear I detected nothing at all at any part of the test.
In my frequency test on $7 \frac{1}{2}$ in. I found that the range of response was as specified, though from input to output (i.e. through the tape twice, once in recording and once in replay) the response was falling above 9 kc . and below about 100 cps . The amplifier has, for playback but not for recording, both bass and treble controls. The treble control can either attenuate or boost ; the bass control only attenuates. In playback I found it best to have full bass response and a slight treble boost.
At these settings with my external loudspeaker system the response was very satisfactory, and even with the internal speaker it was quite good. Indeed I found myself more than once listening with delight to the reproduced music rather than getting on with my testing. The makers' claim that the reproduction is "equal to orchestral recording" is fully borne out. For speech, of course, it is practically perfect.
For a medium-priced recorder I consider this instrument to be wonderful value, and I can thoroughly recommend it.


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reLays - AIRCRAFT INSTRUMENTS

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Supplied as D.C. moving coil, A.C. rectifier and H.F. thermocouple types; also A.C./D.C. moving iron types. Four sizes are available with scale lengths of 2.5 in ., 3.2 in ., 4.2 in . and 6.25 in .

Front of panel or back of panel mounting may be adopted as desired, and if the former method is used there is complete interchangeability with existing round models. The 3.2 in . and 4.2 in . scale instraments are available with either illuminated or non-illuminated dials; the 2.5 in . and 6.25 in . scale instruments being available only with non-illuminated dials.

## WESTON

## Measuring Instruments

## SANGAMO WESTON LIMITED

Enfield, Middlesex. Tel: Enfield 3434 ( 6 lines) \& 1242 ( 6 lines) Grams: Sanwest, Enfield
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 Newcastle 26867 Leeds, Leeds 30867 Liverpool, Central 0230 Wolverhampton, Wolverhampton 21912 Nottingham, Nottingham 42403 . Bristol, Bristol 21781. Southampton, Soton 3328 Brighton, Brighton 28497

## Tune in with

RECD.

## COIL FORMERS




#### Abstract

Wind your coils on "Delanco" S.R.B.P. formers, and improve the performance of your set. "Delanco" formers have been developed after years of research in this field, and in practically every well-known radio or T.V. set, "Delanco" formers are giving trouble-free service. We have a large number of standard types, but maybe your Designs Department wants something new ? Contact us, giving details, and we will produce their "brain-child" with the minimum of trouble or delay.


## ANGLO-AMERICAN VULCANIZED FIBRE CO. LTD.

Cayton Works, Bath Street, London, E.C.I
CLE 327 I.

# VHF TLLEVISION AND THE RECDIVER I.f. 

WITH the possible introduction of another band for Television in this country, bringing with it several alternative programmes, the choice of receiver I.F. becomes even more vital than it now is.

For Band 1 reception, it has been known for many years that the receiver I.F. must be clear of amateur transmitter frequencies to avoid, interference from these transmissions.

The receiver should also be inherently clear of internally generated whistles and patterns arising from beats between the signal and harmonics of the I.F. and local oscillator. The practice of re-tuning the I.F. for different channels, or inserting tunable filters to avoid patterns that would otherwise occur, reduces the flexibility of the superheterodyne principle, and is likely to lead to difficulty should the owner move, or additional stations come into operation.

At the time of the erection of the Sutton Coldfield Station, a $34 \mathrm{Mc} / \mathrm{s}$ I.F. was decided upon by "His Master's Voice" Engineers as the best means of overcoming all the difficulties at that time envisaged, two of which are indicated above. With a choice of programmes, two other problems arise. Their solutions underline the wisdom of the original decision.

Firstly, the local oseillator of the television receiver must not radiate on the channel of another television station in the same area. This has already occurred in the Brighton district, where receivers tuned to Truleigh Hill on $56.75 \mathrm{Mc} / \mathrm{s}$ and employing a low oscillator and I.F. of $11.12 \mathrm{Mc} / \mathrm{s}$ created havoc on receivers still tuned to $45 \mathrm{Mc} / \mathrm{s}$.

By very careful screening, and by use of filters it might be possible on Band 1 to reduce oscillator radiation to such proportions that only in fringe areas will serious trouble arise. On Band 3, however, the efficiency of the R.F. valve as a buffer to prevent radiation is seriously reduced due to the low reactance of the stray and inter-electrode capacities.

Furthermore, aerial filters are of no avail as they, would effectively remove other transmissions to which the receiver may be tuned. The $34 \mathrm{Mc} / \mathrm{s}$ I.F. gives complete freedom from radiating oscillator interference over at least the seven adjacent channels
which it is hoped will be made available for T/V. And hence while it is highly desirable that television designers should reduce oscillator radiation to a minimum, if a $34 \mathrm{Mc} / \mathrm{s}$ I.F. is chosen, at least the residual radiation will not put a neighbour's receiver out of action. The $34 \mathrm{Mc} / \mathrm{s}$ I.F. receiver may well be styled then a "Good Neighbour Receiver."

A further difficulty arises from the fact that the band-width of the aerial circuit at $200 \mathrm{Mc} / \mathrm{s}$ is likely to be of the order of $8 \mathrm{Mc} / \mathrm{s}$, and hence the rejection of the second channel interference on receivers employing a low I.F. is likely to be poor.

It is self evident that with $68 \mathrm{Mc} / \mathrm{s}$ separation second channel selectivity is easily obtained with $34 \mathrm{Mc} / \mathrm{s}$ I.F.

These problems must be of intense concern to the Television receiver manufacturer and to the dealer interested in his future reputation, because although the customer may initially have no trouble in receiving the first Band 3 programme, buying a receiver in good faith, when an alternative network comes into being, possibly of greater power, his reception may be considerably marred by second channel interference from that network or oscillator radiation from a neighbour's set also tuned to that network.

## Adaptors and Converters

Many existing receivers will require an adaptor to receive Band 3 transmissions, and it is obvious that as little modification as possible to the receiver is desirable.

From the considerations indicated above if an existing receiver employs a $34 \mathrm{Mc} / \mathrm{s}$ I.F., it will not only be more easily modified, but will be better placed to receive interference free transmissions, and not to disturb other receivers.

ALL THE "HIS MASTER'S VOICE" superheterodyne receivers manufactured since the opening of the Sution Coldfleld station have employed the $34 \mathrm{Mc} / \mathrm{s}$ I.F.


## SHORTAGE OF RADIO and T/V ENGINEERS

## I YEAR COURSE

We offer full-time day course for one year in the Principles and Practice of Radio and Television.
Next course commences 1st April, 1954.

## 3 YEAR COURSE

This course in Telecommunication Engineering includes one year's Factory attachment. Next course commences 30th August, 1954. brochures giving full details of courses to:

## E.M.I. INSTITUTES

DEPT. I6R, 10 PEMBRIDGE SQUARE, LONDON, W. 2. Telephone: Bayswater 5131/2.
The College associated with a world-wide electronics industry.

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etc.



RA 13 REFLEX SPEAKER
Idea! for critical installations where clear speech reproduction is of first importance. Gives three or four times the coverage of conventional cabinet speakers. Excellent as a local intensifier. Handles up to. 3 watts.


The world-famous range of Truvox Public Address loudspeakers includes many models designed for widely varying applications. Bue all have in common the clarity of reproduction, absolute dependability and magnificent performance under the most exacting conditions which are characteristic of Truvox loudspeakers. The model illustrated is just one example from an infinitely varied range. Write to-day for deseriptive folder and price list.

## ROLA CELESTION LTD.

[^13]'Phone: Emberbrook 3402-6.

# LEAK equipment is unique 

It is acceptable to professional communications engineers for recording and broadcasting. The B.B.C. use several hundreds of the TL/ 12 Amplifier, and $\mathbf{1 , 0 0 0}$ are used by other Broadcasting Corporations.


The Vari-Slope


Representing a unique feedback circuit development, the "Vari-Slope " pre-amplifier gives audibly better reproduction. This advance consists of variable slope "electronic" low-pass filters operating on negative voltage feedback principles. No Inductors ("Chokes ") are used, and their disadvantages are completely eliminated. The turnover frequencies are $5 \mathrm{kc} / \mathrm{s}, 7 \mathrm{kc} / \mathrm{s}$, and $9 \mathrm{kc} / \mathrm{s}$, and the slopes of attenuation are continuously variable over the range 5 db to 50 db per octave.
The filters consist essentially of (b) Extremely low harmonic and inter Twin-T resistor-capacity networks inserted in the return circuit of a single-loop feedback amplifier. The more obvious advantages of this electronic feedback method over conventional choke filters include :-
(a) Improved transient response character. istics (due to absence of chokes having self-capacitance) and the consequent reduction of "ringing."
Frequency amplitude curves for the "TREBLE-3" position ( $5 \mathrm{kc} / \mathrm{s}$ turn-over). Curves of the same slopes are obtained on the other two positions turning over at $7 \mathrm{Kc} / \mathrm{s}$ and $9 \mathrm{Kc} / \mathrm{s}$ ("-2" and "-1" positions).

## Point-One TL/12

## Triple Loop Feedback Amplifier

Used with the " Vari-Slope " pre-amplifier and the best available complementary equipment, the TL/12 power amplifier gives to the music-lover a quality of reproduction unsurpassed by any equipment at any price.
For laboratory use as a stabilised-gain audio frequency power amplifier. For the highest possible standard of disc recording. For the highest possible quality of reproduction from Pickup, Radio, Microphone, Film and Magnetic Tape. For use as a driver amplifier in the speech modulator chain of broadcast transmitters. 27 Gns.

The "Point-One" TL/I2 Amplifier is built to a tropical specification and used throughout the world, including:
The British Broadeasting Corporation.
The South African Broadcasting Corporation.
The Swedish Broadcasting Cor poration.
The Swiss 8roadcasting Corpora-
Theltalian Broadcasting Corporztion. modulation distortion due to negative voltage feedback action.
(c) No discontinuities in the rates of slope when the slope control is operated, and no change in slgnal level at frequencies below turnover (Both these faults occur in variableslope choke fllters due to the slope control altering the terminating impedance and the insertion loss.)
(d) No chokes to cause magnetic hum pickup.
(e) Smaller size, lighte: weight, greater uniformity in production

LIST PRICE $\mathbb{N}$ britain 12 Gns.


## Steep-Cutting Filter

For use with the TL/12 power amplifier and pre-amplifiers preceding the Vari-Slope. This filter unit is of particular interest to the record enthusiast.

Write for fully descriptive literature.

## ELECTRONIC PRECISION EQUIPMENT LTD

## 226 FLWARESECNT LIOHTINE <br> Complete

 kit comprises Hi-craft 40 watt control unit, starter lamp, lamp holders, clips starter lamp, lamp holders, cips tube, 22/6, plus 1/6 post. With tube, $22 / 6$, plus $1 / 6$ post.tube, $30 /$, plus $3 / 6$ carriage.

NEW 5 AMP. THERMOSTAT (MINIATURE)

$2 \mathrm{k} \times 1 \times 1 \ddagger \mathrm{in}$. high
Useful for the control of appliances such as convectors, gluepots, vulcanisers, hot plates, etc. This thermostat is adjustable to operate over the temperature range $50-550$ deg. $F$, fitted with heavy ( 5 amp . A.C.) silver contacts size $1 \frac{1}{2}$ in. long $\times \frac{3}{2}$ in. wide, price $8 / 6$, post 6d.; 1 amp . type, 3/6, 2 amp . type, 5/6.

## FLEXIBLE COUPLINGS

 sometimes
known as bellows couplings because they will extend as well as hend. They are ideal for joining shafts which are out of alignment and for slug tuning controls where the core has to come in and out. Price $1 / 9$ each.


RE-
MOTE CON$\mathrm{T}_{\mathrm{W}} \mathrm{R} O \mathrm{~L}$ only one pair of wires a $n d a$ simple push butcan yelect any one of four stations withour leaving your armchair. This is just one of the many applications just one of the many applications
of our impulse relay. There are of our impulse relay. There are many other purposes to which it can be put. Note they are somewhat soiled, due to storage but mechanically O.K. Price 2/6.


TELEPHONE JACK PLUGS As illustrated 7d. each. Sockets to suit, 10d. each


## VOLUME CONTROLS

We carry a full range of standard size volume controls from 2 K to 2 meg. Prices are: less switch, 2 meg. Prices are: less switch; $3 /-;$, Single pole switch, $4 /-$;
double pole switch, $5 /-$. We can also supply midget-type controls, also supply midget-type controls,
less switch, $4 /-$ single pole less switch, 4/-; single pole switch, 5/9; double pole switch, 6/6. Each of these midget controls has a serial number and carries a 12 -month guarantee by the makers ; they are made on the new moulded track principle and really do perform well.

## A RADIO UNIT FOR SUPERIOR 15

A circuit for a suitable radio unit to fit into our Coronation Console Cabinet has now been completed and throughly tested. All the parts are available. The total cost is E5/19/6. Data is included free with orders for parts, or can be supplied separately; Price $2 / 6$.
Note: This radio unit incorporates T.V. control and is also highly T.V. control and is also highly suitable for fitting into other tele-
visors. The addition of a radio unit visors. The addition of a radio unit
to a televisor is not only worthwhile to a televisor is not only worthwhile
but is essential where the televisor but is essential where the televisor
is kept in a room away from the main radio.
The Superior 15 Corner Cabinet is
 also available now in light oak, or medium oak to suit your furnishings, and it really does look impressive. The price is $£ 18$, plus carriage. H.P. terms $£ 6$ deposit, balance over twelve months

About the Superior 15 itself, if you have not already ordered your set of parts for this, be advised and do so immediately. We are definitely getting down to the last batch of the 15 in tubes and once these are gone the Super 15 T.V. cannot be repeated. At £37/10/- for all the parts (including 15 in. Cossor Tube) this represents the finest value ever offered to the home constructor. If you doubt your ability to make it then send $7 / 6$ for the data and study this first. Don't forget, we guarantee to help you to get perfect results and, if necessary, for nominal charge, will take in your televisor, completely check ove your work, and return it to you in perfect order.

## MAKING TRANSISTORS

Six wire-ended, glass sealed Germanium Diodes. Manufactured approximately two years ago which should therefore be quite approximately two years ago which should thererore be quite
suitable for making transistors, as per last month's article. Offered suitable for making transistors, as per
at the very low price of $10 /$ - the six.
T.V. SIGNAL AND PATTERN GENERATOR Cost of all components, valves, etc., only 29/6. Although this generator can be built and used by any beginner it is at the same time a most useful instrument for the more advanced worker.
It can be tuned to the vision channel and will produce a pattern on the face of the C.R. tube Alternatively if tuned to the sound channel it will produce an audible signal in the loudspeaker.
Thus, its owner will become in dependent of B.B.C. transmissions and can fault-find or test at any time. It operates entirely from A.C. mains and is quite suitable for use with superhet or straigh receivers.
A complete kit of parts (in fac
 everything except the cabinet with full constructional and operational data will be supplied for 29/6, plus $2 / 6$ post and insurance, alternatively, data is available separately, price $2 / 6$ (credited if you buy the kit later)
NOTE. Cabinets as per the illustrated prototype are available price $15 /-$, plus $1 / 6$ post and packing.


## SCAN COILS

Perfectly made by a very famous maker for standard type tubes, we have a limited number only, the price is $12 / 6$ and cannot be repeated once these are cleared, so please act quickly.

## THE ELPREQ E.H.T. GENERATOR

This is a made up unit working on the blocking oscillator/overwound amplifying stage principle. It is of moderate power consumption ( 6.3 volt .8 amp . flament and approx. 59 mA . H.T.) and contains three of the latest BVA all glass valves. Output obtainable ranges from 6 kv . to 9 kv . with normal H.T. rail input but somewhat higher outputs can be obtained with higher H.T. supply. Valve rectification is employed in the output stage. The core of the over wound transformer is kept insulated from chassis in order to give additional protection against possible flashovers. This unit is particularly suitable for those who during the course of their T.V. construcfion have come to appreciate the advantages of separate E.H.T. supply but should also prove invaluable as experimental or stand-by high voltage source. Effective screening (not shown in the illustration) is provided, and the dimensions are $6 \frac{1}{4} \times 4 \times 7 \mathrm{in}$. Price 69/6, post, packing, etc., 5/-.

CRYSTAL SET


Two waves, uses latest crystal valve-good results without bat teries or mains $18 \%$ Headphones, high resistance, $15 / 6$ per pair.
REILAYS


12 CELL ACCUMULATOR
This accumulator can be coupled un to give 24 v , with all cells connected in series or 12,6 or 2 volts in series parallel arrangements. They were originally made for the Admiralty by a eading manufacturer, have never been filled, and are in excellent condition. Each is contained in a wooden crate as illustrated. To clear 10/- each. Post and insurance $2 / 6$

PYREX AER1AL
INSULATORS
Ideal for aerial connections through cabin wall or through panels. Consists of glass dome with threaded rod and terminal ends, and metal fixing flang Price 2/- each.


VIBRATOR UNIT
This unit gives 150 v . at 50 mA from 4 or 6 v. car battery, also gives L.T. supply, suitable for al dry valves. IT4, IR5, etc Ex-W.D. Price $15 /-$, plus $2 / 6$ post.

## LAMP HOLDERS

Bakelite, $1 /$ - each or $10 / 6 \mathrm{doz}$ Bakelite skirted Batten holder, $1 / 6$ or 15/- doz.
Bakelite type threaded for F in with HO skirt, $1 / 6$.
10 per cent. discount if bought in dozens.


CARBON RESISTORS These will now be supplied in individual packets, with the value and wattage cated.

## Prices

2-watt, 5d, each; 1 watt, 6d, each.
L.P. RECORD PLAYER

Made by Decca. Contains B.S.R. motor and Decca pick-up. Although we offer this at approximately half price, it carries same manufacturer's guarantee as if the correct price of $£ 9 / 9 /-$ were paid. Price £4/19/6, plus 7/6 postage and insurance.

GRAMOPHONE UNIT BSR 3-speed motor Type MU14 with Chancery pick-up and two crystal heads, one for long playing and one for normal records. £7/7/-plus $3 / 6$ carriage and insurance.

AMERICAN POWER PACK Built to operate H.R.O. receiver from British mains. Gives $80-100 \mathrm{~m} / \mathrm{a}$. smoothed D.C. at 200 volts, also 5 volts and 6,3 volts. Totally enclosed in metal case. An extremely nice unit. Price £2/17/6, plus 5/postage and insurance.


TOGGLE SWITCHES
Metal body standard size, made by a leading maker. Available with round dolly or with special $\mathbf{V}$ cut dolly. State which type when ordering. Price while stocks last only $2 / 3$.

## MOVING COIL METER

$0-5 \mathrm{~m} / \mathrm{a}$. 2in. flush mounting, Bakelite cased, can easily be made into multi-range test meter. Price $7 / 6$ plus $6 d$. postage.

## MINIATURE 7-PIN PLUG <br> \section*{AND}

Withnon-
breakable
plastic rubber shroud size 1 in. long, diameter approximately ${ }_{3}$ in. Price $2 /$ - pair.

CO-AXIAL CABLE
70 to 80 ohms for T.V. by one of our leading manufacturers, medium thickness. Price 8d. yd, cut to your length.

## L.F. CHOKES

$50 \mathrm{H} \quad 30 \mathrm{~mA}$.
$6 / 6$
50 H 20 mA .
$6 / 6$
$30 \mathrm{H} \quad 20 \mathrm{~mA}$.
$6 / 6$
$20 \mathrm{H} \quad 10 \mathrm{~mA}$. fully shroud $5 / 6$

 $10 \mathrm{H} \quad 75 \mathrm{~mA}, \quad \because . .$. $10 \mathrm{H} \quad 60 \mathrm{~mA}$, fully shroud $20 / 6$ Please add 1/- post. \& pkg.

## COILS

T.R.F. Long and medium wave, complete with circuit diagram, 5/6 pr. Superhet, long wave 900-2,000 metres, medium wave $185-550$ metres, short wave $16-50$ metres with aerial and oscillator coils that is set of six. Coils with circuit diagram, price 10/6.

GLASS INSULATOR
Complete with iron bracket for mounting on pole, wall, tree, etc. Álso suitable as aerial insulator. Price $3 / 6$ each.

ELECTRONIC PRECISION EQUIPMENT LTD.

## GIVE AWAY PRICE

only
$\delta 7$ or
47/-
deposit.
Less valves and power pack
famous set by a famous manufacturer.


Undoubtedly a serious listener's receiver,
Among many special features are an H.F. stage and tuning indicator, Tunes up to 11 metre band. Price complete with valves but less speaker, $£ 14 / 19 / 6$. H.P. terms $£ 5$ deposit and 12 monthly payments of $£ 1$.
We have a few left, less valves and power pack, otherwise in good condition; they definitely have never been used. Price $£ 6 / 19 / 6$, or $£ 2 / 7 /$ - deposit and 11 monthly payments of $10 / 9$, plus $15 /$-carriage.

## THE INFRAY LAMP

Means real comfort in bed or in workroom or other place where air temperature is low as it emits Infra-Red Rays which not only warm you but relieve pain, if you have any, and keep you healthy.

- Economical because its rays warm you and not the room.
- Costs only $\ddagger \mathrm{d}$. per hour to run (electricity at 1d. per unit).
- Works off lighting circuit (full instructions supplied).

- Absolutely safe for continuous burning, no health or fire risk
Ideal for many other uses:-over pet's basket-rearing pup, chicks-over desk-work bench, etc.
- Completely and unconditionally guaranteed for five years,

All complete and ready to work. Price $36 /=$, post \& pkg. 2/Money refunded in full if after seven days' trial you are not completely satisfied.

We regret that, owing to a typographical error, the price of the Truvox Tape Deck advertised in our February advertisement was given incorrectly. The correct price, of course, should be 22 guineas, or H.P. terms £7/14/- deposit with the balance spread over 12 months.

## BEDROOM-NURSERY MAINS MIDGET RADIO

All 11 th the parts, bakelite cabinet, will cost you only $£ 3 / 15 /$ postage) which, we think you will agree, postage) which, we think you will agree, is not too much to spend on your dear ones. The set is economical to run, too for it uses only three valves in a specia. reflex T.R.F. circuit which gives ample power combined with good tone. Incidentally if you wish to give the sets to young children why not decorate the cabinet with a few suitable transfers? These can usually be obtained from local handicraft shops. Circuiting and construction data free with the parts or available separately at $1 / 6$.
THE ELPREQ "sELECTIVE FEED-BACK" AMPLIFIER The amplifier is fitted with independent bass and treble
 control, both connected through different feed-back loops so that no "cut" at all in the ordinary sense is applied. The variation which can be achieved, by applying various degrees of negative feedback in the higher and lower ranges of the sound strata will accommodate all individual tastes. We stronely.
a 12 in . speaker in recommend the fullest use of the instrument's potentialities, Booklet and set of components available at once at $£ 2 / 19 / 6$, post, etc., $2 / 6$. Booklet separate $1 / 6.12 \mathrm{in}$. speaker to suit $\Sigma 3$, post free if bought with a mplifier.


15in. MAGNETIC TELEVISION TUBE By famous maker, Specification Blue/White screen 9 Kv , ion trap triode, heater 6.3 v . at .55 amp ., $50^{\circ}$ deflection. New, with written guarantee, offered at approximately half price, $13110 /-$ each, plus 10/= carriage and insurance. H.P. terms $£ 4 / 10 /$ - deposit and 12 monthly payments of $18 / 3$. Limited quantity, so order immediately.

SLIDER RESISTORS
Heavy Duty Type.
Size 7 in . $\times 1 \frac{1}{2} \mathrm{in}, 11$ ohms 4.5 amp., 22/-; Size 9 in . $\dot{x}$ $1 \frac{1}{2}$ in. 1.2 ohms .15 amp ., 15/-; size $13 \frac{1}{2} \mathrm{in} . \times 1 \frac{1}{2} \mathrm{in} .3$ olims 10 amp., 15/-.


LECTROSS TOWEL AIRER, £3/19/6 The Lectross warms room as it dries clothes, bathing costumes, towels,
etc. Size 3 ft . wide, 3 ft . high and 5 in . deep. Works of A.C. or D.C. mains, consuming 650 watts. Fully guaranteed. Price £3/19/6 plus $7 / 6$ carriage.

40 VOLT $2 \frac{1}{2}$ AMP. STEP
DOWN TRANSFORMER
Totally enclosed in metal boxprimary 200-240 50 c.p.s. ; secondary easily rewindable to other voltages-ideal safety unit for operating children's toys or domestic gadgets, 17/6 each


CHASSIS ASSEMBLY
3 colour, 3 waveband scale covering standard Long, Medium, and short wavebands, scale pan, chassis punched for standard 5 valve superhet, pulley driving head, springs, etc., to suit. Scale size $14 \frac{1}{12}$ in. $\times 3 \frac{1}{2}$ in. Chassis size, $15 \mathrm{in} . \times 5 \mathrm{in} . \times 2 \mathrm{in}$. deep Price $15 /$-, plus $1 / 6$ post. Note: This is the one that fits our 37/6 table cabinet.


DEMOBBED VALVES
Gives the commercial equivalents of many thousands of Service Valves, an invaluable publication recently revised. Price $2 / 3$.


THE WHITE LADY
Moulded in expensive pure white bakelite, this cabinet is ideal for T.R.F. or Superhet alike. It is supplied with holes for two controls only. Size approx. 11 in . $\times 6$ in. $\times 5 \mathrm{in}$. Price 22/6, plus $3 / 6$ post and ins.


THE WINDSOR DE LUXE This takes our Windsor 5 chassis and 8 in . Speaker, is undoubtedly a fine cabinet, well-made and pleasantly finished. Approximate $8 \frac{1}{2}$ in. Price $49 / 6$, plus $3 / 6$ post and ins.

THE PORTA-RADIO
 This is ultramodern, wo-tone, bakelite with integral moulded handle. We can supply, supply, required, the metal chassis, dial, and all other parts necessary to make a Mains or Battery portable. Note: All of these cabinets have slight imperfections; these are hardly noticeable, however, and will not impair the performance or safety of the set. Price 10/- each, post and ins. $3 / 6$.

THE MINI-RADIO


Internal dimensions of this are $6 \frac{1}{i n} . \times 5 \frac{1}{2} \mathrm{in} . \times 3 \mathrm{in}$. Two models are available, one has the new plastic "open crackle" finish. Price 15/9, plus $1 / 6$ post. The De Luxe model is covered with brown and grey leather cloth. Price 22/6, plus 1/6 post. Either model has fitted carrying handle.

## ELECTRONIC PRECISION EQUIPMENT LTD



## THE PROJECTOR

An impressive cabinet, originally designed for T.V. but slight modification makes it into an unusual, but most dignified, radiogram or amplifier. Size 23 in. wide, 22 in . deep, $37 \frac{1}{2}$ in. high. wide, 22 in . deep; $37 \frac{1}{2}$ in high. Price $\begin{aligned} & \text { C8/riage and ins. } £ 1 \text {. }\end{aligned}$


This is a robust, handsome carrying case, finished in two-tone leather cloth with foldaway carrying handle and clasps. Approximate dimensions $19 \mathrm{in} . \times 14 \frac{1}{2} \mathrm{in} . \times 10 \mathrm{in}$., ample for 8 in . speaker, mains or battery driven amplifier, and tape deck or other record-playing mechanism. Price £3, carr. and ins. 10/-


THE REGINA
High polished, nicely figured, twotone walnut finish, developed for the home constructor and supplied with component parts for a shelf that can be fitted in at any desired height. Suitable for most constructor sets, "Viewmaster,", "Supervisor,"" "Magnevisor," "Teleking, ${ }^{3}$ etc., etc. Size $18 \frac{1}{2} \mathrm{in}$. wide, 18tin. deep, 34in. high. Cut for 12in. tube, but not drilled. Price $£ 7 / 17 / 6$, or $£ 3 / 12 / 6$ deposit.

THE TABLE-TELE
Another cabinet which through the misfortunes of a manufacturer we ate able to offer at below cost. Designed to take a 12 in . tube, we supply this complete with armoured plate glass at $£ 3 / 17 / 6$, moured plate glass at $7 / 6$ post. Metal work, punched chassis, outrigger, etc., available as a parcel. Price 17/6, plus $2 / 6$ post.


THE NON-REPEATABLE
Through a manufacturer ceasing production we are able to offer this really handsome cabinet at well below cost. It was originally made for a very expensive televisor so its quality is beyond question. Size 1 ft .10 in . wide, 1 ft . 4 in . deep and 3 ft . 5 in . high. Complete with plywood back, fitted " Bowler Hat," Price $£ 7 / 5 /$-, or $£ 3 / 8 / 4$ deposit. Note: The cutout is for 12 in . tube but holes for the controls are not drilled.

THE MINI-MIKE
A brown bakelite cabinet with metal grille. Size $4 \frac{i n}{} \times 4 \frac{1}{i n}$. $x$ 2in. Will take 3 hin, moving coil speaker or other type of microphone. Price 6/6, post and packing 1/-



THE MIDGETRONIC
Yet another bankrupt bargain. This pleasing little cabinet (size approx. $8 \mathrm{in} . \times 7 \mathrm{in} . \times 3 \frac{1}{2} \mathrm{in}$.), moulded in bakelite is supplied complete with dial ring and special pointer as illustrated. Price 15/-, plus $2 / 6$ post. Or complete with all the valves and parts to make an excellent T.R.F. set, price $£ 3 / 15 /-$ plus $2 / 6$ post. Note: A few suitable transfers make this an ideal nursery cabinet.


THE OCCASIONAL
Available in cream or brown bakelite, this is probably our most popular cabinet. Approximate size 11 in. $\times 6 \frac{1}{2}$ in. $\times 5 \mathrm{in}$. The price $17 / 6$, plus $3 / 6$ carriage and ins. Or with all parts, including four B.V.A valves, to make a T.R.F. receiver of proved design, T.R.F. receiver of proved design,
price $£ 6 / 1 / 6$, or if req., $£ 2 / 1 / 6$ price $£ 6 / 1 / 6$, or if req., $£ 2 / 1 / 6$
deposit and ten monthly payments of $10 / 6$. Postage and ins. is $3 / 6$.


THE SMALL INSTRUMENT This is of sheet metal finished black crackle with lanule front left ready for lettering with white paint or transfers. Approx. size 8 in. $x 4$ in $\times 6$ in. high. Price $15 /-$, plus $1 / 6$ post and packing.


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Finished French walnut and eucalyptus veneer, highly polished. Twin playing shelves for ampliplaying shelves for ampli-
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Please add postage under £1. C.O.D. or Cash with order. C.O.D. charge extra-Open 9 a.m. $\mathbf{- 6}$ p.m., Monday to Friday. Sorry but we close at $1.30 \mathrm{p.m}$. on Saturday.

## You're SURE to get it at Sterins <br> FOR HOME <br> CONSTRUCTORS <br> A 5 VALVE 3 WAVEBAND SUPERHET RECEIVER for $£ 10 / 10$ - <br> For use on A.C. Mains 200 to 250 volts. The

 following are outstandlag features; - A superhet circuit designed for high - A 3ifin. P.M. speaker accurately- 4 in. P.m. speaker accurately matched

The latest range of new
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Send $2 / 6$ for the fully descriptive stage by atage assembly and wiring diagrama, with
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 (A) FEEDER UNIT. Complete up to and including Audio stage, A.V.C.
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## WIRELESS WORLD 3 VALVE SET

## TWO BATTERY PORTABLES <br> (a) THE "MINI TWO-THREE

An " Alldry" Battery Portable of midget size. 6ifn. $\times 4 \$ 1 \mathrm{n}, \times 3 \% \mathrm{in}$., designed to covet medium waveband $190-559$ metres, with use of short trailer aerial.
The simple design of this Recelveris so arranged that cither a 3 -valve set or a 2 -valve (afterwards ensily converted to the 3-valve) can be made. Consists of a T.R.F. circuit using a regencrative detector
 With H.F. stage and a high

case), and the 3 -valve for filety bullt for $£ 4 / 3 / 6$ (luss includes valves, speaker and $£ 5 / 3 /-$ (less case). Each price


Send 2/- for the assembly
instructions: they include simple and complete practical
component layouts ind diucomponent layouts and diuinexperienced constructor to successfully build either set. All components are arailable for separate sale, a price list being supplied with rassembly
instructions.

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(b) The "MINI-FOUR" A* 4-valve Battery superhet 4 pre-set stations, three on mengum waveband and one en conditions. Each station is obtainad on the set by the turn of a rotary switch. No tuning is It is of midget size, being oniy $4 \mathrm{itn}, \times 8 \mathrm{in}$. $\times 4$ inn. when completely built and is very assily assembled from diagram supplied.
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## AN AMAZING OFFER! A COMPLETELY ASSEMBLED

4 VALVE T.R.F. CHASSIS
Including a 5 in . P.M. SPEAKER and VALVES FOR ONLY \& 6'9'6
This receiver is of the very on A.C. or D.C. Maina. It
 W avehands and ind Medium modern BVA miniature paive The line up being 12 BAf. incore- 12A6-35W4. It
Tuned Colls, thus ensuring
The overall silze of the cond sengitivity.
The overall size of the complete chassis
including speaker 18 101in, $\times 4$ tin. $\times 6$ in.
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The DENCO M.T.O.I. Modulated Test Oscillator £3/15/(Plas 2/-carr, and ins.). Has Frequency range continuously variable from $170-475 \mathrm{Kc} / \mathrm{s}$ and $550-1,600 \mathrm{Kc} / \mathrm{s}$. Battery operated and thereby completely sel fcontained

## "PERSONAL SET" BATTERY ELIMINATOR

A complete Kit of parts to bulld Midget "Alldry" Battery Eliminator, giving approx. 69 volts and 1.4 volts. mains and Is sultable for any 4-vaive Superhet Beceiver requiring H.T. and L.T. voltage
approx to 69 ribove
a
approx.to 69 volta.
The Kitis quito easily and
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In addition we can offer a simllar COMPLETE KIT to provide approx. 90 volte and 1.4 volts. Size of assembled unlt $7 \mathrm{in} . \times 2$ sin. $\times$ litin. Price $47 / 6$.

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Constructors everywhere are amazed!

## tre "TELE-VIEWER"

5 CHANNEL TELEVISOR

## DESIGN OF A COMPLETE 12 " SUPERHET T.V. RECEIVER

"/ HUNDREDS SOLD IN 4 MONTHS "" OERECT FRINGE AREA RECEPTION
(i) SIMPLE DIAGRAMS MAKE :- PERFEG FRINO AKITER REGEPTION AT HALF

Here are some of the features which combine to make this such a fine receiver The Superhet circuit easily tuned to any of the five channels, i.e. LONDON, SUTTON COLDFIELD, HOLME MOSS, WENVOE and KIRK-O-SHOTTS. (The extreme ease of tuning is accomplished by the provislon of pre-aligned I.F.T.s.)

- A lifelike, almost stereoscopic, picture quality made possible by the following factors
a. Excellent band width of I.F. circuits
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c. C.R.T. Grid modulared from low impedance source
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The picture brilliance is also much above the average and enables

- FIMM pable viewing with normal room lighting or daylight.

FIRM picture "HOLD" circuits (Frame-Line) ensures a steady picture, free from bounce or flicker even under the most adverse conditions met with in " fringe"," areas and excellent " interlace " ensures the absence of "liney effect."

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2/3 watts of High Quality Sound.
We can supply a
New 12 in. C.R.T
at specially re-
duced price of

- Rigid C.R.T. mounting enables entire receiver to be safely handled
- All preset controls are mounted on side of chassis enabling all adjustments to be carried out whilst facing the C.R. Tube. As no hire purchase terms are available the receiver can be bouth in five separate stages (practical diagrams and circuits are provided for each stage) thus enabling hire purchase interest rates to be avoided. The complete set of ASSEMBLY INSTRUCTIONS is now available, price 5/-. The instructions include really detailed PRACTICAL AAYOUTS, WIRING DATA AND COMPONENT PRICE LIST, ALL COMPONENTS ARE AVAILABLE FOR INDIVIDUAL


## Entire receiver buift on two chassis units each measuring $14 \frac{1^{\prime \prime}}{2} \times 6 \frac{1^{\prime \prime}}{} \times 3 \frac{t^{\prime \prime}}{}$

PURCHASE. A CABINET WILL ALSO BE AVAILABLE.

## NOW available at Stern's

## The "WIDE ANGLE" TELE-VIEWER

- A design that retains all the distinctive features of the 12 in . Televisor but with increased Time Base ciency, producing 15 to 16 kV . E.H.T., with ample scanning power for C.R. Tubes up to 17 th . $\left\{\begin{array}{l}\text { It can be complotely built in- } \\ \text { clading sapply of all valves tor }\end{array}\right.$ (1) $\left.\begin{array}{l}\text { (plus cost ot C.R.T.) and is as simple } \\ \text { to construct as tbe 12in. model. }\end{array}\right\}$
- Thista the most effecent " WIDE ANGLE" large screen design yet offered to constructors, and yet lt
- Complete assemblyinstructions, dlagram, etc., available for $5 /$ -



## A 4-VALVE QUALITY "o PUSE-PUIL Incorporating Negative Feedbsck. Filter Input Circult and

 employing fitsin Push-Pul.. A simple arrangement is provided be used, ind Is suitable for use with Standard or long-playing reconds. A tone control is incorporated, and the 10 -watt outpur tranaformer is deaggned to match $2 t$ to 15 obm speakers 7 ifn. high, and full practical dlagrans are supplied. Price, incinding drilled chassifund valves, of complete kit, $£ 6 / 17 / 6$. Price of assembled chassis, supplicd ready for use. $£ 8 / 12 / 6$. Plus 5 -Carr. \& Ina. Full descriptlve lenfets are aval ableseparately for 1 -

12-watt EXGH FIDELITY "PUSU-PULL" AMPLIFIER destgned for A.C malne 200 to 250 volts employs 6 valves plus rectifer with negative feedback, and comprises a main amplifer chassle and a remote controlled Preamplitier and Tone Control Unlt, lncorporating four controls-bass, treble, main volume or mixing control, and a radio, gram, ral erophone, selector switch. Thls control anit measures only $7 \times 4 \times 2 \mathrm{in}$. The measured frequency range of the smplifer "ith this unit howa an excellent response from 14,000 cycles down to 20 cyclee. the bass and treble controla allowing independent control of gain at both ands of the frequency range from zero to a gasn of 50 . It can be seen, therefore that ample correction is provided to suil any type of pick-up with any type of recording. Input voltage for maximum output is 70 mV .6 .3 volts at 2 amps , and 30 mA . H.T. Is provided for tuning nuit. etc. Price of complete kit, including drilled chasais and valves, £14. Complete speclfication and layout, $2 /$. We ean also supply ela. . THIS AMPLIFIED COMPARES WELL WTTE THE WHIAMSON AND SIMILAR DESIGNS AT A FRACTION OF THEIR COST.

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All Ktts are for A.C. Mains 200-2000 Volts. They comprise as Metal Rectifier and Transiormer, Lapped Selector 12 volt charging, and a tapped Resigtor, with
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For 6 or
F For 6 or 12 vold batceries max. 1 amp. An eanly followed wiring Diarram is included with An eashly followed Wiring Diagram is included with

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 SUPERHET COIL TURRETS WITH A ROTARY TURRET ACTION Type CT9 consists of as four-station "pre-set." unit fromwhich any three statious on medium waveband and one on long wave can be, recelved by a turn of the turret switch Price 39/6.
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$111^{\prime \prime} \times 43^{*} \times 3 \xi^{\prime \prime}$
A desigm of a complete 5 -VALVE SUPERHET RE CEIVER employing an $\mathbb{R}$. $\mathcal{P}$. Stage and incorporating a separate VIBRATOR PACK size $41 \times 21 \times 6$ ind. for
use on 6 or 12 volt D .0 . supplies. We can supply all components to build this complete Recelver and Vibrator Pack Including a Metal Case

Valves. Drilled Chass18 and $5 \ln$. P. M. Speaker. Of the $£ 12 / 19 / 6$. (Carr. and Ins. $5 / 6$ extra)

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employing new Componenta.
Aend $2 / 8$ for the complete
STRUCTIONS, CTRCDITS and of ARSRMBLY IN. OUTS, including a complete Individual Component Price List.

## Modernise youn old Ractiognam

## THREE COMPLETELY ASSEMBLED

 ALL-WAVE SUPERHET CHASSISModel B.3. A 5-valve 3-waveband Receiver
Model B.3.P.P. A 6-valve 3-waveband Receiver with PUSH-PULL OUTPUT,

- Model B.3.P.P./R.F. A'7-valve 3-waveband Receiver
incorporating an R.F. stage with PUSH-PULL OUTPUT.
The three Receivers are for operation on A.C. mains $100 / 200$ volts and designed to the most modern specification great attention having been given to the quality of reproduction which gives excellent clarity of speech and music on both gram. and ratio, making them the Ideat replacement chassis for that "old Radlogram," etc.
Brief specifications: Model B.3.Brief specifications: Model B.3.-
Valve line-up, 6BE6, 6BA6, BAT
Valve line-up, 0BE6, 6BA6, 6A
6BW6, 6X4-waveband coverage short ${ }^{\text {6 }}$ 16-50 medinm maveband coverage long $900-2,000$ metres. trols: (1) volume
with on/off; (2) with on/off; (2)
tuning (fywheel
type); (3) ware-
change and gram. (4) tone (3-pasition switch operative on gram. and radio) is employed over the entire audio stages. Chassis
size: $11 \times 7$ ? $\times x$
size high. Dlal
Price complete and READY FOR USE, excluding spenker, £12/12/(carr. and Ins. 7/6 extra).
Model B.3. P.P. This model is the B. 3 Receiver butincorporates two 6BiV6 VALVES in PUSH-PULL, resulting in really excellent quality
reproduction up to approximately 6 watts. Price $£ 15 / 15 /-$ (plus $7 / 6$ reproduction up to approximately 6 watts. Price $£ 15 / 15 /$ - (plus $7 / 6$ Model 8 Ins. Model B.3. P.P./R.F. This modelis simillarin appearance and has same waveband coverage as the Model B.3, but in addition it incorporates an
R.F. STAGE together with PUSM-PULL OUTPUT, employing a total of 7 valves rith two type 6BW6 in Push-Pull. This makeal for a really sengitive receiver with genulne quality reproduction. Price £18/18/(plus 7/6 carc. and ins.)


A dealgn of a simple 1 -vaive 2 -stage Battery Receiver, giving excellent results on medium and long wavebands and having exceptionslly law battery consumption. Drilled chassis and practical diagrams make it the idea
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PLESSEY, 10 in . 3 ohm V/coll
ROLA, 12in. 3 ohm V/coil.
BAKERS, 12 in .15 ohm V/coil
$\begin{array}{lrr}\text { £1 } & 5 & 0 \\ \text { c2 } & 9 & 6 \\ \text { £3 } & 19 & 6\end{array}$
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## WE ALSO HAVE THE NEW W. B."STENTORIAN"

HIGH FIDELITY SPEAKERS IN STOCK
Model H.F. 6-inch.
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 quallity reproduction for the lower-price range. 3 or 15 ohm models are available

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 We have had very considerable experience in assisting customers to butd this T/V and can gupply sPECl instructions showing practical layouts and price list Holme Moss, Kirk-o-shotts and Wenvoe.This AUTOCHANGE UNIT by a
Famous Manufacturer is offered for $\mathbf{5 1 1 / 1 4 / 6}$

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(Normal price is $816 / 10$ /-- These units will auto change on all three speeds $7 \mathrm{n} ., 10 \mathrm{in}$. and 12 in . - They play MIXED 7 in Ther have. records. - They have separate supr.p.m., which are moved Into f.p.m., which are moved Into - Minimum baseboard siza gight above bulk purchase enables us to ofter these BRAND NEi UNITS at this exceptional price
The COLLARO 3RC/521 3-Specd AUTO CHANGE UNIT $\mathcal{L 9} / 17 / 6$
$\left\{\begin{array}{l}\text { We will supply this } 3 \\ \text { speed Autochanger } \\ \text { and the Model B. } \\ \text { Chassis on the left } \\ \text { together with a loin. } \\ \text { (or 8in. P.M. speaker } \\ \text { for £23 or with the } \\ \text { B.3.P.P. for f26/5/- } \\ \text { or with the Model } \\ \text { B.3.P.P./R.F. for f29 } \\ \text { Carr. and Ins. } 7 / 6 \\ \text { extra. }\end{array}\right.$
(Including Carro and Ins.).

With
heads for standard and $\mathbf{L} \mathbf{P}$ heads for standard and L.P. records.
weight adjusting
Will antochange 10 in . and 12 im . records not intermixed.

- Minimum Base plate size 16in. $\times 12 \mu \mathrm{n} .$, with heigh above 4yin. and below baseplate $3 \ln$



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In addition to the Record Players advertised we have a
FEW ONLY - NEW MANUFACTURER'S SURPLUS COLLARO MODEL 3RC/522 LATEST COLLARO 3-SPEED AUTOCEANGER This is the same type of unit as the 3RC/J21
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Incorporating the fmous CONNOISsEUR Light Weight Moving Iron Head and including the Connoie seur matching Transformer (plus 1/- carriagr $39 / 6$
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DUAL-CHANNELPRE-AMPLIFIER and TONE CONTROL UNIT This comprehensive PRE-AMPIFIER
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A-speed non-miker AUTOCHANGE UNIT complete
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switch on the pilc-ap head.
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RIM DRIVE 78 rip. pm complete with $t$. Collaro Pluy Hype MAGNETLC TURNTABLE. Thes BRAND NEW COMPITE or A.C.Mail is 200-25il Volts.

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SHADED POLE RIM DRIVE (0/* GRAM MOTORS
(Plus 1/- carr, and tns.) Panel. Could also be used as Recording Take Џp or Rewis Motor

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HALF-WAVE H.T. RECTIFIERS
250 Volta $150 \mathrm{~m} / \mathrm{a}$
250
200
Vols 250 m

## SELENIUM

 RECTIFIERS6 or 12 Volt 1 amp. rating 7/6 or 12 Volt 24 amp rating $12 / 6$ 6 or 12 Volt 6 amp rating 1 alif/9

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TII54B TRANSMITTER UNITS


Medium/high powered for C.W.-M.C.W. R/T, 3 ranges $10-5.5 \mathrm{Mc} / \mathrm{s}$. $5.5-3 \mathrm{Mc} / \mathrm{s}$. $\quad 500-200 \mathrm{kc} / \mathrm{s}$. Complete with 4 valves, etc., in metal case 14 in . $x$ $16 \frac{1}{2}$ in. $x 8 \frac{1}{2} \mathrm{in}$. External Power Supply required. ASK FOR $39 / 6$ each $\quad$ CARRIAGE
X/ESA EXTRA ALSO AVAILABLE.

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15/= each CARRIAGE
CARRIAGE
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## MONITOR CRYSTAL TYPE 2, 10T/11390

## As used with the R1II6 or R1082.

Less valves and Crystals, but otherwise complete, dim. $7 \frac{1}{2} \mathrm{in}, \times 5 \frac{1}{2} \mathrm{in} . \times 3 \frac{1}{2} \mathrm{in}$. Plastic constructions, in transit case.
ASK FOR
5/= each 1/- EXTRA
MAGNETIC MARCHING COMPASS MK. I
Small hand compass. Can be used day or night. Complete with instructions, in plastic case, size $3 \mathrm{in}, \times 2 \frac{1}{2} \mathrm{in}, \times \frac{7}{6} \mathrm{in}$.

| ASK FOR |  |  |
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| X/H406. | $12 / 6$ | each POST |

## INDICATOR UNIT TYPE 6

In original wood case.
Containing VCR. 97 tube and valves, etc., in metal case $18 \mathrm{in}, \times 8 \frac{1}{2} \mathrm{in} . \times 7 \frac{1}{2} \mathrm{in}$. New condition.
ASK FOR $\quad 79 / 6$ each CARRIAGE
X/H524. PAID
MIDGET MOTOR, Ref. 5 U/2705

Input 24 v. D.C. 2 a., R.P.M. 2,800 drive pulley each end. Overali dim. $2 \mathrm{in} . \times 2 \mathrm{in} . \times 5 \frac{1}{\frac{1}{2} \mathrm{in}}$. | ASK | $7 / 6$ each | POST |
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| $\times / H 98$. |  |  |

## MALLORY SYNCHRONOUS

Input: 12 v. D.C. Output approx. 120 v. D.C. 70 mA , unsmoothed, dim. $5 \frac{1}{2} \mathrm{in} . \times 2 \frac{3}{4} \mathrm{in} . \times 5 \mathrm{in}$. Unused, but soiled and vibrator contacts stuck due to long storage.
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10/= each POST

## MICROPHONE REF. 10A/14381

(Flying Helmet Type.)
Electro Magnetic 500 ohms with switch, lead and 2 way sockets.
ASK FOR
3/11

| ASK. FOR |
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| X/EI6. |
| $26 / 11$ each $\left.\begin{array}{l}\text { POST } \\ \text { 6d. EXTRA }\end{array}\right)$ |

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## 26 WATT OUTPUT TRANSFORMER

Parmeko type AF5084/IA. (Mfg. Surplus). Primary: 6,600 ohms. C.T. Sec's. 3.5, 5, 7.5 or 10 ohms. Dim. $3 \frac{7}{8} \mathrm{in}, \times 2 \frac{3}{8} \mathrm{in}, \times 3 \mathrm{in}$. Fully shrouded Weight 3 lbs.
ASK FOR
19/6
each

| POST |
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## CABLE

2 core Metal braided $X / H 541$, 4d. per yard, post 3d. extra.

HALF MILE REELS (880 YARDS)
23 SWG P.V.C. covered signal corps wire, X/H855, 25/- per reel.
CERAMIC WAFER ROTARY WAVE.
CHANGE SWITCH

3 bank each, single pole, 5 ways. Overall length 8 in. $\times 2$ tin, wide.
ASK FOR
3/- each 3d. PACH

## VALVEHOLDERS

Ceramic B9G (EF50), British Octal (SP61), Int. Octal (5Z4), 7UX (RK34)
1/- each, post I $\frac{1}{2} d$, extra. Dozen lots $10 / 6$.
Moulded, British Octal, B7G (6AM6) Moulded. British Octal, B7G (6AM6), B8A (EAF42), B8G (ILN5). Int. Octal, B9A (I2AU7), 7UX (6Aㄱ).
9 d. each, post $1 \frac{1}{2}$ d. extra. Dozen lots $7 / 6$. Paxolin, B9G (EF50), B9, B7G, B8G.
6 d . each, post $1 \frac{1}{2} \mathrm{~d}$. extra. Dozen lots 51 -.

## CONDENSERS

Mansbridge Metal cased type, dim. $5 \frac{1}{2} \mathrm{in} . \times 4 \mathrm{in} . \times$
1 ỉin. XRN. Capacity 6 mid. wkg. vig. 1,000 volts.

5/6 each 9d. EXTRA
X/H359.
Electrolytic Tubular. Aluminium cased, wad.
cardboard cover. U.S.A. made, Capacity 32 mfd. wkg. vtg. 450 v. D.C. Dia. 2 in ., length $4 \frac{1}{2} \mathrm{in}$. with mtg, plate.
ASK FOR
ASK FOR
$1 / 9$ each 9d EXTRA
Or 3 Condensers $(32 / 450)$ as above, in original sealed carton.
ASK FOR
4/6
POST
Electrolytic Tubular
Capacity $8-16 \mathrm{mfd}$. wkg. vig. 450 v . D.C. Dia.

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$3 /=$ each
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Electrolytic Tubular. Aluminium case. Capacity 8 mfd. wkg. vtg. 450 v. D.C. Dia, lin. $x 2 \frac{1}{2} \mathrm{in}$. Tag ends. Mfg. surplus.
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$\begin{array}{lll}\text { ASK FOR } \\ \text { X/H598. } & 1 / 9 \text { each } & \text { 3d. EXTRA }\end{array}$
POST
Ceramic Tubular. Mfg. Surplus. Capacity PF $1,100,1,000,504,200,180,150,120,82,68$, $56,39,33,15,12,8.2,6.8$. Price bd. each, post $1 \frac{1}{2} d$. extra. Dozen lors 5/- post paid.
Silver Mica. ( $w$ ) waxed. ( $p$ ) pitched. Mfg. Surplus. Capacity PF 5,000 (w), 4,550 (w.), 4,000 (w), 432 (w), 432 (p), 350 (p). 135 (p), 47 (w), 20 (w), $10(w), 8$ (e), 5 (w). Price $6 d$. each, post $1 \frac{1}{2} d$. extra. Dozen lots $5 /$-post paid. Tubular Paper (pitch covered) 400 V. D.C. wkg.
Mig. surplus. Capacity mid. $0.068,0.039,0.022$. $\begin{array}{llll}\text { Mig. surplus. } & \text { Capacity mid. } & 0.068, & 0.039, \\ 0.018, & 0.022, \\ 0.015, & 0.0082, & 0.00068, & 0.0056, \\ 0.0039,\end{array}$ $0.0027,0.0022,0.0018,0.00082,0.00068$. Price 4d. each, post $1 \frac{1}{2} \mathrm{~d}$. extra. Dozen lots $3 / 6$ post paid.
Tubular Paper (pitch covered) 600 V. D.C. wkg. Capacity mid. $0.012,0.0082,0.0033,0.00082$. Price 6d. each, post $1 \frac{1}{2} d$. extra. Dozen lots 5/post paid.
VOLUME CONTROLS. Potentiometers, Carbon Track.
$\frac{1}{2}$ meg. ( 500,000 ohms.), long spindle fitted DP/ST Switch and mtg. plate.
ASK FOR POST
X/E189. 3/11 each 3d. EXTRA
With DP Switch Ref. $3 \mathrm{U} / 100 \mathrm{M}$. PAC $\mathrm{X} / \mathrm{H} 415,500,000$ ohms., ref. $3 \mathrm{U} / 500 \mathrm{M}$ PAC $\mathrm{X} / \mathrm{H} 655$. Price $3 / 9$ each, post 3d. extra.
With SP Switch 3in. spinde, 2 meg. ohm. ref. 187-Z-2169.
ASK FOR
X/H427.
3/3 each. 3d. EXTRA
Less Switch. 3in. spindle $\frac{1}{4}$ meg. ( 250,000 ohms).
ASK FOR $2 / 3$ POST
$\times / H 705$.
X/H705. $2 / 3$ each 3d. EXTRA
Preser Types, less switch.
(M) Metal type, (T) Tropicalised, 10,000 ohms. Ref. $10 \mathrm{~W} / 8855, \mathrm{X} / \mathrm{H} 434$ (T), 25,000 ohms. Ref.
$10 \mathrm{~W} / 9230 \times / \mathrm{H} 433$ (T) 100,000 . IOW/9230 X/H433 (T) 100,000 ohms. Ref.
IOW/7457 X/H414 ${ }^{(M)} 200,000$ ohms. IOW/7457 X/H414 (M) 200,000 ohms. X/H956 (M). 250,000 ohms. Ref. $10 \mathrm{~W} / 9188 \times / \mathrm{H} 431$ (T). Price $1 / 6$ each, post 3d, extra.

Short Spindle Types, less switch
50,000 ohms. Ref. $10 \mathrm{~W} / 8741$ tin. $5 \mathrm{p} . \mathrm{X} / \mathrm{E924}(\mathrm{~T})$ 100,000 ahms. Rel. $10 \mathrm{~W} / 8777 \pm$ in. sp. X/H699 (T). 500,000 ohms. Ref. $10 \mathrm{~W} / 7856$ in. ipp . X/H700 (T). 500,000 ohms. Ref. $10 W / 7856 \frac{1}{2} \mathrm{in}$. sp. X/H733 (M)
1 meg. ohm. Ref. ZA. 2809 . $\frac{1}{2} \mathrm{in}$. sp. X/H73 1 meg. ohm. Ref. ZA. 2809 , $\frac{1}{2} \mathrm{n}$.
Price $/ / 9$ each, post 3 d . extra.
Midget Types, léss switch.
500 ohms. Ref. $10 \mathrm{~W} / 8990$. $\frac{1}{2}$ in. sp. X/H428. 25,000 ohms. Ref. $10 W / 17403$, Preset, $X / \mathrm{H} 657$. 100,000 ohms. Ițin. sp. ............ X/E928. 250,000 ohms. Linear. 客in. sp. ...... X/H896. 500,000 ohms. Log, I\%̄in. sp. ........ X/H895. Price $2 / 3$ each, post 3 d . extra.
W.W. POTENTIOMETERS

2 watt by Clarostat U.S.A. 10,000 ohms. Ref. CMC-63532 Preset.
ASK FOR
$2 / 6$ each
POST
X/H957.
3d. EXTRA
3 Watt Type.
500 ohms. Ref. DB1307/12. 弪in. sp., $X / H 661$. 1,000 ohms. Ref. $10 \mathrm{~W} / 8924$. Preset. X/E936. 2,000 ohms. Ref. $10 \mathrm{~W} / 8843$. $\frac{1}{4} \mathrm{in}$. sp., X/H904. 5,000 ohms. Ref. low/7796. in. sp., X/H423. 5,000 ohms. Ref. $10 \mathrm{C} / 8925$. $\frac{1}{2}$ in. sp., X/H663. 10,000 ohms. Ref. 10C/8926. Preset, X/E568. 20,000 ohms. Ref. CLR 4037/17S. Iin. sp., X/H421. 25,000 ohms. Ref. $10 \mathrm{C} / 8927$. Preset, $\mathrm{X} / \mathrm{H} 665$. 25,000 ohms. Ref. $10 \mathrm{C} / 538$. $\frac{1}{2}$ in. sp., $\times / \mathrm{H} 422$. 25,000 ohms. Ref. $10 W / 8573$. tin. sp., $\times / H 522 A$. 25,000 ohms. Ref.
50,000 ohms. Ref. CLR $4001 / \mathrm{gS}$. Preser, X/E943. 50,000 ohms. Ref. CLR4001/8
Price $3 /$ - each, post 3d, extra.
5 watt, 2 hole fixing type. 270 ohms. Ref. CLR/6003/I2. W9259. lin. Spindle. ASK FOR

2/- each 3d.EXTRA
X/H905.
2/- each
J. EXTRA

5 watt type. Ref. $10 \mathrm{~W} / 8534$. $\frac{1}{2 i n}$. sp., X/H57I.
500 ohms. Ref. $10 H / 6524$. $\frac{1}{2} \mathrm{in}$. sp., X/H666.

25,000 ohms. Ref. IOW/8572. $\frac{1}{2}$ in. sp., X/H667. Price 3/- each, post 3d. extra.
10 watt Toroidal Type.
200 ohms. Ref. Wi363, wound on porcelain.
ASK FOR 3/= each 6d POST
X/EI72. $3 /=$ each 6d. EXTRA
FIXED RESISTORS, Wire Wound.
25 watt, 26 ohms. Vitreous finish, wire ends, Ref. $10 \mathrm{~W} / 15688$. Type AW3192.
Ref. each, post 3d, extra. Dozen lots 30/- post paid.
18 watt, 2,000 ohms. Vitreous finish Contact ends.
2/- each, post 3d. extra. Dozen lots $19 / 6$ post paid.
15 watt 50 ohms. Vitreous finish. Tag ends.
2/-each, post 3d. extra. Dozen lots 19/6 post paid.
10 watt Vitreous finish Wire ends. Resistances, ohms: 6,800, 270,25,10.
2/6 each, post 3d. extra. Dozen lots 25/-post paid. 5 watts, Cemented (C), Enamelled (E), Resistance, ohms: 1,200 (E), 220 (C).
$2 /$-each, post 3d. extra. Dozen lots $21 /$ post paid. 4 watt 100 ohms. Vitreous finish Wire ends.
2/- each, post 3d. extra. Dozen lots $19 / 6$ post paid. 3 watts Enamelled Wire Ends, Resistance, ohms: $2,200,1,800,1,500,1,000,150,33,18$.
9d. each, post 3d. extra. Dozen lots $7 / 6$ post paid. 1.5 watts, Enamelled Wire Ends. Resistance, ohms $1,500,1,200,1.000,200,1.7$.
4d. each, post $\frac{1}{2} \mathrm{~d}$. extra. Dozen lots $3 / 6$ post paid. Resistance type 849, IOW/767. Heavy Duty type W.W. or bakelite mrg. Panel $5 \frac{7 i n}{2} \mathrm{in}$ x $2 \frac{1}{2} \mathrm{in}$. $\times$ $3 / 16 \mathrm{in} .0 .3$ ohms, tapped every 0.05 ohms.
ASK FORT
POST $\times 1 \mathrm{H} 686$. $3 /=$ each 6d. EXTRA
$\frac{\text { Dozen lots } 30 / \text {-post paid. }}{\text { RESISTANCE TYPE } 243}$
$10 \mathrm{~W} / 9305 \mathrm{~W} . \mathrm{W}$. on mica panel $43 \times 18-\mathrm{mm}$, with fixing holders. 1.3 ohms.
ASK FOR $3 /=\quad$ POST
X/H687. $3 /=$ each 6d. EXTRA
Dozen Lots 30/- post paid

## RESISTANCE ELEMENT TYPE 4773

10W/16125.
Heavy Duty. I ohm. element NO former or mtg.
ASK FOR 3/ POST X/H688. 3/- each 6d. EXTRA Dozen lots 30/-post paid.

RESISTANCE TYPE 150 10W/8524
W.W. bobbin type on former 10.5 cm . long, 18 mm . dia. with fxg. holders. 40,000 ohms. tapped every 5,000 ohms.
ASK FOR $3 / 6$ each 3 POST
$\times / H 689$.
Dozen lots $32 / 6$ post paid

## Order direct from:

## CAR RADIO SPECIAL_Partly assembled



Small size case $12 \times 4 \times$ 6 in. Will fit most cars. For either 6 or 12 volts, depending on vibrator. Chassis supplied with 5 octal valve holders, medium wave aerial and oscillator coils, outand orams ormer volume pur trol sundry resist conces and condensers, dial and kobs Case dial and knobs. Case finished in brown crackle. Dial calibrated 150-550 metres. 5 valves to suit. One each, either GT or metal: 6SA7, 6R7, 6V6, 6K7, OZ4.

LASKY'S PR1CE $55 / 5 /-$ Carriage $5 /$ - extra
Or less valves, 69/6. Carriage 5/- extra.
Other chassis in various conditions of completion are available for personal callers only.
CIRCUIT for 5 valve car radio, using above chassis. PRICE 1/6.


## R.1132.A RECEIVERS

Supplied in maker's original wood transit case Frequency coverage $100-124 \mathrm{Mc} / \mathrm{s} .11$ valves 1 VR65, 1 VR66, 4 VR53, 2 VR54, 1 6J5, 1 VS70, 1 VR57. Large tuning scale with slow motion drive, $0-5 \mathrm{~mA}$. tuning meter. R.F. and L.F gain controls, jack sockets for line and phone Totally enclosed in metal case, grey enamelled with plated handles. Size: $18 \times 10 \times 11$ in Supplied with all valves, also circuit and calibration chart.
GRADE 1. Brand New, 79/6. GRADE 2. Soiled, 49/6. GRADE 3. Secondhand, 39/6. Carriage 10/- per unit extra.


RESISTANCE AND CAPACITY BRIDGE
For A.C. mains 200/250 volts. Complete with valve rectifier and 6 H 6 and EM34 (magic eye) valves. Ranges: Ohms Factor of 0.1 to 10. Farads. In metal case, black crackle finish, $12 \times 6 \times 8$ inches. Without handles. This unit is ideal fo breaking down and rebuilding as another type of instrument.

| LASKY'S PRICE |
| :--- | :--- |
| Carriage $3 / 6$ extra. |



GARRARD RECORD PLAYERS For 6 and 12 v. operation, complete with magnetic pick-up and volume control. In metal cabinet size: $17 \mathrm{in} . ~ X$ 14 in . $\times 11 \mathrm{in}$. Very limited quantity. LASKY'S PRICE
£5.19.6

## R. 1155 RECEIVERS BRAND NEW AEBALL TESTED

 These well-known ex-Air Ministry Receivers need no further introduction. Supplied complete with 10 valves, and full circuit data. LASKY'SPRICE
USED
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Only. £11.19.6 Covering the shipping band of 1.5 $3.0 \mathrm{Mc} / \mathrm{s}$. Price $£ 17$.
Carriage 12/6 per unit extra, including 10/- returnable on packing case. 10 s . 0d. rebate will be given on power packs for the R. 1155 when purchased


Fully Assembled Power Pack and Output Stage, for R1155 Receiver. For use on 200-250 volts A.C. mains. LASKY'S PRICE
Carriage 5/- extra. $\quad \mathbf{~}$
with the receiver. 5d. each, 4/- doz.

PF.,
12 VOLT D.C. MOTOP GENERATORS
Output 300 volts at 150 D.C. 7,500 r.p.m. diam. 6in. long. LASKY'S PRICE

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$465 \mathrm{Kc} / \mathrm{s}$ Iron dust cores in cans, midget type. Size 1 in . $x$ lin. $\times 2 \frac{1}{2}$ in. By Plessey. Price 8/6 per pair. Kc/s. 8/6 per pair. $50 . \quad 450-470$ WEARITE
Kc/s. $8 / 6$ per pair.

## BAFFLE RADIO CABINETS



Pleasing design, complete with knobs, drilled chassis, dial, drum drive and back. Finish in satin mahogany veneer, natural colour polish. Outside dimensions: $17 \frac{1}{2}$ in wide, $11 \frac{1}{\text { in }}$. high, 5 in . deep.
Circuit and component list for valve T.R.F. set available.

LASKY'S PRICE
$36^{\prime} 6$
Carriage 2!- extra.
A LASKY'S RADIO ADVERTISEMENT SEE OVER.

nous. For $200-250 \mathrm{y}$ 50 c.p.s. Many uses. LASKY'S
PRICE

- / 6

METAL
RECTIFIERS
6 and 12 volt F.W. 2a Bridge.

2a.
$9 / 11$
12/-
$17 / 6$

## HEARING AIDS <br> By well-known Manufacturer. In metal case, size: $2 \frac{1}{\mathrm{i}} \mathrm{in} . \times 4$ in in . $\times 1 \mathrm{in}$. Complete with batteries and 3 sub-miniature valves earpiece and cord. Ony two controls Fitted with interna crystal microphone.

suitable for reconstruction into mildget
LASKY'S PRICE $99 / 6$
Postage 3/6 extra. 99/6
Ready for use. Pertect working order. slightly soiled, but new and unused. A few hearing aids vailable. lens earpiect, coril and batterie

TANNOY PRESSURE UNITS 10 watts. 7.5 ohms impedance. Last few only. PRICE
Carriage 4/6 extra.
59/6
TWO-GANG TUNING CONDENSERS . 0005 MFD .
No. 1. Miniature. With trimmers. Size : $24 \mathrm{in}, \times 11 \mathrm{in} . \times 11 \mathrm{in}$. tin. spindle. With trimmers. Size: $\quad$ LASKY'S PRICE 6/6
 No. 3. Midget. Less trimmers. Size: $2 i n . \times 1 / 2 \mathrm{n} . \times 1 \frac{1}{2}$., spindle. Standard LASKY'S PRICE $6 / 6$ No. 4. Standard type. Size: $2 \frac{1}{2} \mathrm{in} . \times 2 \mathrm{in} . \times 1 \frac{3}{3} \mathrm{in}$. tin . spindleTHREE GANG. 0005 mfd . Less trimmers. Size: $3 \frac{1}{2} \times 23$ x $1 \%$ in. LASKY'S PRICE 6/6.


LINE TRANSFORMERS FOR "ETRONIC"
T.V. RECEIVERS

No. 1. For models 1536 and 1637. Complete with EY51 rectifier, 39/6.
No. 2. 7 Kv . type, $35 /$-.
POT/METERS. All values. Wire Wound from $\mathbf{3 / 6}$. Depending on wattage and length of spindle. Carbon. Less switch $2 / 11$ each With s.p. switch .... 4/3 each With d.p. switch .... 5/6 each VCR97 C.R. TUBES, new unused. 35/-. Carriage 5/-.
Screen Enlarger for VCR97. Filter or clear, 17/6. Postage $2 / 6$. C.R.T. Neck Protectors, $2 / 6$.

10 K.V. METROSIL E.F.T. REGULATORS. By Metrovick. Pencil type, 5/- each.

## S.T.C. SENTERCELL RECTIFIERS

RM.1. . $3 / 10$ K $3 / 40,3.2 \mathrm{kV} .6 /-$ | RM.2 . . $4 / 3 \quad \mathrm{~K} 3 / 50,4.0 \mathrm{kV.8/8}$ |
| :--- | :--- | RM.3.. 5/$\mathrm{K} 3 / 100,8.0 \mathrm{kV}$ 14/8

RM.4. . 18/-
$\mathrm{K} 3 / 160,12.8 \mathrm{kV}$.
6- AND 12-VOLT
VIBRATORS
4-Pin type. Soiled. S/H. New

State voltage required.
8-PIN JONES SOCKETS. For
1155 Receiver, etc., $1 / 9$ each
WESTINGHOUSE
$250^{\circ}$ v. R.M.S.STIFIERS
14A976. $80 \mathrm{~m} / \mathrm{a}$.
14A86. $200 \mathrm{~m} / \mathrm{a}$
14A100. 270 v. R.M.S.
$200 \mathrm{~m} / \mathrm{a}$.
TRIPLEX DARK SCREEN FILTERS
$14 \times 121 \times \frac{3}{3} \mathrm{in}$.
$15!\times 13+{ }^{\frac{3}{3}} \mathrm{in}$.
$15 \frac{1}{2} \times 134 \times \frac{3}{18} \mathrm{in}$. $\ldots \ldots . .$.
Postage and packing $5 /-$ per piece extra. (This charge is necessary owing to extra packing required.)

DARK SCREEN PERSPEX FILTERS $14 \mathrm{if} . \times 14 \mathrm{in} . \times \ldots . .$. $13 \mathrm{in} \times 1 \mathrm{in}$

PERSPEX. $13 \frac{1}{2} \mathrm{in} . \times 10$ in. $\times$ fin. Neutral shade slightly marked, $4 / 11$ per piece.

BRANDENBURG R.T. E.H.T: UNITS
$6-9 \mathrm{Kv}$ Complete with valves. $19 . . .{ }^{6} 19$
13-16 Kv ........... $£ 8196$
TEST PRODS
Fully fused, with retractable points, $4 / 11$ per pair ( $1 \mathrm{red}, 1$ black).

## TOGGLE SWITCHES.

BULGIN.
S.P.S.T.
D.P. Change over

25/14/11

## SPECIAL C.R.T

## OFFER

Brand new and unused 12 in . ion trap cathode ray tubes. 6.3 volt heater, 7-9 Kv. E.H.T. 35 mm , neck. Black and white picture. By famous white picture.
manufacturer. PERFECT ${ }^{\text {P1 }} 2 / 19 / 6$
Carriage and insurance Carriage and insu.
$15 /$-per tube extra.
MANUFACTURERS SURPLUS T.V. COM PONENTS
Wide Angle Scanning Coils. Low imp. line and frame $\ldots$ pair $19 / 6$ Scanning Coils. 35 mm. Low imp. line and frame
Frame output transformer Standard Focus Coil. 35 mm . electro magnetic
Line or Frame B.O.
transformer. Auto.
Wide Angle Frame B.O. trans.
P.M. Focus Magnets.

With vernier. 35 mm .
Tetrode
Triode
Wide Angle P.M.
Focus Unit. For all 38
mm. tubes. With
vernier and picture shift, Ferroxdure.

## PLESSEY

Scan soils per pair 25/Width Control ...... 6/6 P.M. Focus magnet $12 / 6$

Co-Axial Cable. 70-80
ohms impedance.
Single core, 8/- doz. yards, Twin core, 12/- doz, yards. Twin feeder, 6/-doz, yards. Co-Axial Connectors. For standard in. cable, 1/6.
WX6. WESTINGHOUSE MINIATURE
RECTIFIERS Wire ends. 1/6 each
C.R.T. MASKS

Brand New
LATEST ASPECT
RATIO
10in.
12 in . Flat Face
12 in . Old ratio
14in. Rectangular .... $9 / 6$
15in. Creamgular. . . . $12 / 6$
15 in With fitted
safety glass .......
16in. Plastic, white
16in. Double D
17in. Rectangular

## PLASTIC

ESCUTCHEON
SAFETY MASKS
Incorporating dark screen
12 in . Round Face 12in. Double D. Round Face .

12/6
16in. for metal tubes $25 /-$
SOIL.ED. NEW ASPECT RATIO MASKS

## 9 in .

12in. . with ${ }^{12 \mathrm{in}}$. fied
12 in . with fitted
armour plate glass, cream plate glass, $11 / 6$
ARMOUR PLATE GLASS
16in. Actual size 171
$\times 15 \frac{1}{2} \times$ inch
15 in. Actual size $168 \mathrm{in} . \times 13 \mathrm{in} . \times 1 \mathrm{in}$. 12 in . Actual size 13 in . $6 \times 10$ in. $\times 4$ in. $2 / 6$ 9in. Actual size 9in. $3 / 6$ 8in. $x$ tin

## DE LUXE T.V. CABINENS

## Our new 12 inch model. Mark II

This cabinet is now supplied complete with mask, glass, castors, shelf, bearers, c.r.t. neck end protector, back, speaker fret and baffle board. Finished in beautiful figured medium, iight or dark walnut veneer, with high polish. Suitable for most home constructor T.V. receivers, including the "Viewmaster," "Practical Television," "T Tele King," "Magniview, , supplied with cut-out for Can be suppe no extra cost.

An allowance of 4 s .6 d . will be made if the mask is not required.
Mask and glass extra when cabinet is ordered with cut out for 14,16 , or 17in. C.R.T.'s.
Inside Dimensions : Depth 161 in .; width 17 in. ; height 28 in . Overall height 32 in . and width $18 \frac{1}{2}$ in.
WHY NOT CONVERT YOUR TABLE RECEIVER TO A CONSOLE MODEL.
Adaptor frames for fitting 9 in . or loin. c.r. tubes can be supplied if required.


## 

## THE VIEW MASTER

Construction envelope $7 / 6$. POST FREE Wide Angle Conversion 3/6. POST FREE
All components in stock. Write for price list
R.F. E.H.T.

OSC. COILS For use with 6 V 6 Heater winding for EY51. Circuit and full data supplied. 6-10 Kv. PRICE 19/6 $6-18 \mathrm{kV}$.

PRICE 25/-
COL.ARO 3-SPEED AUTOMATIC RECORD CHANGERS


MODEL 3RC/521.

Brand new and unused in maker's original carton.
Pleasing cream or fawn finish. Complete with hi-fidelity studio crystal turnover head. Type GP. 29.

## LASKY'S PRICE <br> Carriage Free.

## £9.19.6

THE "UNIVERSAL" LARGE SCREEN AC/DC TELEVISOR
By A. S. Torrance, A.M.I.P.R.E., A.M.T.S.
A 28-page booklet giving full instructions for building a large
17 -inch screen televisor.


3d. POST FREE.

TYPE AT/9
T.V. MAINS

AUTO-
TRANS-
FORMER
and 375 volt
tappings 250 mA . Also 5 v mA . Also $5 v$. 3 a.; 6.3 v . 3 a.; secondaries 3 a. secondari
Price 25/-.
ION TRAPS
All types. Price
3/-. State tube type number when ordering.

## INTERCOM UNITS

4-station operation. For use on A.C./D.C. mains $200-250$ volts. Supplied complete, with 3 new valves, ready for immediate installation. Fitted in attractive plastic cabinet.
Suitable for use as baby alarm. MASTER UNIT $£ 5 / 19 / 6$.
Carr. 5/-extra.
Extension Units. Price 21/- each complete. Carriage $2 /$ - each extra.

## LASKY'S LINE

RF.EHT for line flyback. $6-8 \mathrm{Kv}$, with EY51 heater winding. Suitable for home construction T/V able for h.
$19 / 6$ each.

POSTAGE STAMP TRIMMERS
Paxolin. Up to 100 pF . 6d. each. 5/- per doz. Ceramic. Up to 100 pf . 9 d . each. $7 / 6$ per doz.

Duodecal (B12A) bases. VCR139 c.r.t. bases. 1/- each. 10/6 dozen.

ELAC DUOMAG FOCALISERS. For Wide angle c.r. tubes.
Low, medium and Low, medium and

## THETELE-KING A practical 5-channel SUPERHET TELEVSION RECEIVER Using the new 16 and 17 inch cathode ray tubes and wide angle components for the home constructor. <br> Complete instructions, wiring diagrams and 32-page descriptive booklet <br> 6/= POST FREE <br> ALL COMPONENTS IN STOCK <br> WRITE FOR LIST

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| :---: | :---: |
| ANGLE COM. | Power pack |
| PONENTS | Sound-vision and |
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| type Ferroxcube | PRICE 11/- each. |
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| Frame B.O. | surplus. $83 / 16 /-$ |
| transformer 15/- | COILS 13 all ex- |
| Line EHT. | actly as specified. |
| transformer 40/- | Price 44/6. |

## OUTPUT TRANSFORMERS

40 mA Multi ratio
80 mA Multi ratio
80 mA Pentode
00 mA Pentode ........ 1216
60 mA Plessey, $6,000 \mathrm{ohms}$
Standard Pentode
Pentode
Midget Pentode
Miniature Pentode, 3 S4, is 4 PX4 Intervalve 5:1 Intervalve HEAD SLOW MOTION
JACK PLUGS AND SOCKETS
Standard size $3 / 11$ per pair.

RESISTANCES. 72 Resistances,
18/-.
CABINET
As illustrated here. $\mathbf{~} 8 / 10 /$ Carriage $12 / 6$ extra.

## WIDE ANGLE CATHODE

 RAY TUBES14in. MW36-22 14 in . C14B $\quad \cdots . . . \sum_{2} 10 \quad 1$ 16in. MW41-1 …. $\mathrm{E}_{22} 410$ 16in. T901 17in. MW43-64 17in. C17BM Carriage and insif $£ 13 \quad 1$

## P.M. LOUDSPEAKERS

All with 3 ohm speech coll
 3 ln . $14 / 6$ 5in. $14 / 681 \mathrm{in}$. $15 /$ NEW AVAILABLE 12 -inch Goodmans heavy duty speaker. Capacity 15 watts, 15 ohms peech coil impedance.
LASKY'S PRICE $£ 5 / 19 / 6$. Car. $3 / 6$ ex. All loudsprs. offered are first grade and of highest quality construction. Many other types in stock. Send us your reqts, types in stock. Send us your reqts, Special offer.
ohms. LASKY'S PRICE $49 / 6$.

## LASKY'S T.V. CONSTRUCTORS' PAR CELS

No. 1. All brand new components by Igranic. Comprises E.H.T. flyback line transformer, $7-10 \mathrm{Kv}$. with ferroxcube core and rectifier heater winding: scanning coils; frame output transformer; Elac focus unit with vernier adjuster, U37 E.H.T. rectifier and brand new 12-inch cathode ray tube with ion trap, mask and glass. LASKY'S PRICE FOR THE COMPLETE PARCEI, $£ 15 / 19 / 6$. Carriage and insurance, 15/extra.
No. 2. The Constructors ${ }^{\text {2 }}$ Parcel as above, but less the cathode ray tube and ion trap. LASKY'S PRICE 79/6. Carr. 3/6 extra.
No. 3. Condenser Parcel. 1 of each: $-04 \mathrm{mfd} .12 .5 \mathrm{Kv} . ; 32$ +32 mfd .350 v.W.; $32+100 \mathrm{mfd}$ 450 v.w. AND $241,000 \mathrm{pf}$. ceramic tubes; ${ }^{6}$. Imfd. 500 v.w. .01 mfd . 500 . V.w. ALSO 12 assorted "pf" condensers of your own choice. PRICE 45/-. POST FREE.


No. 4. Completeset of metalwork, as illustrated here. Unassembled. Comprising main chassis, tube supports and valveholders. (Less sound-vision chassis.) PRICE 25/-. Carriage $3 / 6$ extra.
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NOTICE TO ALL PURCRASERS OF THE ENGLISH ELECTRIO 16 Inch The C.R.T. TYPE T. 901
service. By only reconditioning reconditioned English Electric. A 19 and carries maker's tube cost antee. Write for turther detrils

MAINS TRANSFORMERS
All 200-250 v. 50 c.p.s. primary Finest quality, fully guaranteed. M.B.A. 3 350-0-350 v. 80 mA ., 6.3 v. 4 a., 5 v. 2 a. Both filaments tapped at 4 volts. An ideal replacement trans.

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MBA/6. $\quad 325-0-325 \quad \mathrm{v} .100 \mathrm{~mA}$. 6.3 v. 3 a., 5 v. 2 a. With mains tapping board. Price 22/6.
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Price 18/-.
MBA/8, 235-0-235 v. 60 mA . and 6.3 v. 3 a. Price 12/6. AT/3. Auto transformer. $0-10-$ $120,200-230-240$ volts 100 watt. Price 17/6.

## Expanded Metal. Silver Finish

$12 \mathrm{in} . \times 12 \mathrm{in}$. ............... $3 / 11$
$12 \mathrm{in} . \times 18 \mathrm{in} . . . . . . .$.
Plastic, White, $12 \mathrm{in} \times 5$ in. 2/-
SMOOTHING CHOKES
20 mA .40 H
40 mA .8 H
40 mA .10 H
$100 \mathrm{~mA} .10-20$
250 mA .10 H

## J/RA/3

6 VALVES 12-15WATTS OUTPUT
Originally made for talkie film projectors. In carrying case. Chassis size $14 \frac{1}{2} \times 10 \times 4$ in For use on $100-250$ volts A.C. mains. Grey crackle finish. Fitted with volume and tone controls. Fitted with volume and tone controls. Resistance capacity coupled cull
Output 2 KT 66 in push-pull.


High quality components used throughout. in black rexine covered wood case, size $15 \frac{1}{2} \times 13 \frac{1}{2} \times 9 \frac{1}{2}$ in. giving plenty of room for speaker, etc. Circuit diagram available. Complete with 6 valves, fully assembled
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R.F. OSC. COLL KITS Consisting of R.F. oscillator E.H.T. coil with EY51 heater winding, EY51 rectifier, 6V6 valve and base. All necessary condensers and resistances. Including full circuit and data.
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 6BE6, 6BA6, 6AT6, 6BW6, 6X4, Flywheal tuning. Negative feedback over entire audio sectio
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-Douhle Feature with P/Pull
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Model B6. Output 6 wat
4 Short. (3 Band spread). Long,
4 Short. (3 Band spread)
5 valves. Oufput $3 t$ w.
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215150
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Output 6 watt
PULLY GUARANTEED.
$\begin{array}{ll}223 & 2 \\ T\end{array}$ For A.O. Mains 100/120 and 200/250 volis. Tho Double Feature ehastis are supplied with - separate

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Primaries tapped $200 / 250 \quad \vee$.
Type MT5B. To chs rge 12,6 , and 2 V . at $2 \frac{1}{2}$ amps, $18 /-$ Type MT5. To charge 12, , and 2 v , at 4 amps. $22 / \%$. Plus $1 / 6$ post and packing.

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 Conslsting of 2 specially designed chokes and 3 condensersExtremely effective, cuts out all mains noise. Can be Extremely effective, cuts out all mains noise. Can be assembled in existing receiver or separately as desired.
Complete with circult diagram, $4 / 11$, plus 1/-P.C.

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D.O.,2in.sq., panel mountlog, 16/6, plus $1 / \cdot$ post and packing.

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6.3 v . If amp., $5 / 11: 6.3$ v. 8 amp., $8 / 11 ; 6.3$ v. (with
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Sllver Mica; Moulded Mica; Wax Tubular; Paper
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$8 \mathrm{mid} .450 \mathrm{~m}_{8}$.
$8-8 \mathrm{midd}, 450 \mathrm{~F}$.
$8-16 \mathrm{mfd} ., 450 \mathrm{v}$.
$16 \mathrm{mfd} ., 450 \mathrm{r}$.
$32 \mathrm{mfd} ., 450$ ₹.
$50 \mathrm{mfd} ., 13$
v.

RESISTORS (CARBON)
${ }^{\frac{1}{2}} \frac{1}{2}$ and it watt, 3d. each. 1 watt, 5 d . each.
LARGE STOCKS OF B,V.A. VALVES and Er-Govt. Special purpose valves.

Four enguiries invited.

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$\mathrm{K} 3 / 50.4 \mathrm{kV}$.
$\mathrm{K} 3 / 100$.
8.5 KV.
METAL RECTIFIERS-BRAND NEW !
$300 \nabla .75 \mathrm{~mA}$, mayy be used In series or voltage doubling
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Ill the above plus $1 / 6 \mathrm{pkg} .$, earr., ins.
GOODMANS 8in., H.F. 15 obms 5 w. peak A.O.
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W.B. 6 ir . B.F., 10,000 lines, 3 ohms
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WRARFEDALE BRONZE 10in. Flux
Density, 10,000 lines $6 \mathrm{w}, 15$ ohms...........
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 Suspension, Flux Density 13,000 lines, 40 w. WHARFEDALE SUPER 5/Cg/AL Sin. Cloth suspension, Flux Density 18,000 lines, 10 ohms
WHARFEDAKE Wi2/Cs 12in. Cloth Suspension. Flux denalty 13,000 lines, 10 \%.
WHARFEDAEE Wi2 1210. Flux Density
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15 W. Peak A.O. ............................
12
6

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Type LP1. $250.0-250,60 \mathrm{~mA} .6 .3 \mathrm{~V} .3 \mathrm{mmp}, 5 \mathrm{~F}$

- LP4. $\quad 275-0-278 \quad \nabla$. Rit $80 \mathrm{~mA} ., 0 / 4 / 6.3 \quad$. at

LP6. $\quad 350-0-350,80 \mathrm{~mA}, 6.3 \mathrm{v} .5 \mathrm{amp}$, , 5 v
3 amp, Halfshrouded
LP6. Fully shrouded as above
LP7. $350-0.350120 \mathrm{~mA}, 6.3 \mathrm{v}$.
4 amp. Fully shrouded $\quad . . . \ldots$.....
LP9. $\frac{425-0-425}{} 150 \mathrm{~mA} .3$ v. 4 armp. C/T
V .3 smp. Fully shroude
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voltage
adjustment panel tapped $\begin{array}{lll}\text { roltage adjustment panel tapped } \\ 200 / 210 & \nabla_{1}, 220 / 230 & \nabla, 0 \\ 240 / 250 \\ \nabla\end{array}$
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B14. 2 amp. Universal upright mounting.
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B15. Generai Purpose step-down TransGorcher. Tapped $3 \nabla \cdot 4 \%, 5 \nabla ., 6 \mathrm{~F}$., 24 ₹., 30 v. Total output 30 ₹. at 2 amps.

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$8412 \quad 8$
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Acos Mic. 22 (Crystal).
LUSTRAPEONE
M.C. With T/F.C51

Heavy Table Base for above RESLO (M/C Low Imped.)

Rlbbon High Fidellty
Mumetal Transformer $\qquad$
$\begin{array}{lll}84 & 4 & 0 \\ 8 & 0\end{array}$ $\begin{array}{rrr}25 & 15 & 6 \\ 21 & 1 & 0\end{array}$ $\begin{array}{lrr}57 & 5 & 0 \\ 815 & 0\end{array}$

## THE NEW ACOS 33-1

Crystal Mike,, Ontput level-55 db. Ref. 1 จ./dyne/cm:, 50/- pluc 6d, post/pkg

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A high impedance crystal malcrophone complete with
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Incorporating high impedance mu-metal twin-track heads. Tro-speed capstan, for tape speeds of $7 \frac{1}{4}$ and 34 lnches per and rewind facilitiea without tape handling, All controls operated by electrically and mecbanlcally interlocked puah buttons.

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TAPE SPEED : 7 HLn sec, FOR USE ON A.C. $200 / 250$. 50 CYCLES MAINS ONLY

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FERROGRAPH, $1,200 \mathrm{ft}$.
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THE NEW COLLARO STUDIO PICK-UP
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Valveline up: 12J7, 35L6, 1487, receiver, lacorporates s carrive receiver, incorporates a carring transportable. Housed in an attractive plastio cablinet of modern denign, it cars be used on either A.C. or D.0.200/250 v AVATLABLE IN ONR COLOUR ONLT: WALNUT
BROWN. BROWN. GUARANTBED FOR TWELTE MONTES TYANA" SOLDERING IRON Weight 4 ozs.i Adjustable Bit; heating time 3 mins. Length 101 in. 14 , 1 plus $1 /$ - post. GOLDER GUN Instant heating, low voltage, bit whlch can be bent to shape
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MODEL I T.R.F. RECEIVER
Thisis a 3 valve plus metal rectifer TRF recelver with a valve line-up as follows: 8K7 (HF), $6 J 7$ (Det) and bVe (Output). The dial lsilluminated and when assembled the receiver presents a very attrictive appearance. Coverage sfor the Medium and Long Wave bands. Operates on $200 /$ 250 volts AC Mains.
Plus $2 / 6$ Packing,
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Thisis a powerfui midget 4 valve plus metal rectifier Superhet Recelver uith a valveline-up asfollow: : $6 \mathrm{~K} 8,6 \mathrm{GK}, \mathrm{BQ}, 7$, 6V6. The dialistlluminated and coveraze is ior the short
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£7.19.6
T.R.F. RECEIVER We can supply this Recelver ready buill at $£ 619 \mathrm{~s}$, 6 d , plus $3 / 6$ p.e. ALL COMPONENTS SUPPLIED ARE GUARANTEED FOR ONE YEAR
NOTE: We would reapectfully sugyent to those interested in building this recciter that they send for OUR Instruction Booklel. Intending eonstructors can then judge for THE M SELVES how comprehensive this Boo ivet is.
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make high quality tape recordings of live speech or music, gramophone or radio and telephone conversation, etc.

- Dual speed $3 \frac{3}{4} / 7 \frac{1}{2} i n$. per sec.
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The 'Impresario' can also be used as a high quality radio, gramophone or microphone amplifier.

## PRICE $4 \pm 1 / 2$ GNS.

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GUARANTEED
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The ' Impresario 'Is the first transportabie tape recorder in Great Britain to provide power supply and Internal space for a Radio Tuner unit with optional listening and/or recording
DISTORTIONLESS SUPERHET PRICE T GNS May be fitted in 3-WAVE RADIO TUNER UNIT (Tax Paid) 14 GNS. a few minutes.
The 'Lmpresario' combined instrument will not only provide tape recordings from microphone, gramophone, radio, telephone, etc., but can be used lndependently as a high-quality ampliter. It also provides internal space with H.T. and L.T. supplies for a radio tuner unit.
a straight ampllfer when recording playhack is supplemented by a neutral position which enables the unit to be operated as a straight ampliner when recording facilities are not required. An internalinput terminalis provided for connection to a becomes ar radio recelver. Should it be required to record any part of a radio programme, it \& necessiary only to switch to record " and atart the tape deck.
Thein put jack for" mike," gram, etc., automatically mutes the radio. The speaker jack is so arranged that any programmes may be monitored or an extension speaker used if required. A system is employed whereby the Internal speaker may be used as an occasional microphone for recording.
The e W ampliderincorporate egenerousnegative feedback over each of Its Ave atages. The use of independert bass and treble
 The antest Truvox 3 -motor, dual-gpeed ( $37 / 7 \mathrm{inn}$. per second) $t$ win track, electrically and mechanically Interlocked pushbutton control, Tape Deck is used, and the Impresario is housed in an attractive light brown imitation lizard skin cabinet, measuring 19 by $13 \frac{1}{2}$ by 133 in . Migh. It 1 s for operation on $200-2507$. A.C.
The radio tuner unit can be fitted very easily ; the right alde panel of the cabinet ls conveniently detachable, being held by fourscrews. The tuner, is a modifled version of the Elplco RF/716 3 -way superhet unit,

## Ammonncing a NEW 4 watt AMPLIFIER KIT

This is a 3 ralve 3 atage Amplifier for use with Gramo phone, Microphone or Radlo. Vulve line-up is ás follows
68L7; 6V8: 5Z4. Negative feed-hack. Tone control. Volt68L7; $6 \mathrm{~V} 8: 524$. Negative feed-back. Tone control. Volt-
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comprehensive point-to-point winng circuit diagramo comprehensive point-to-point winng circuit diagran
Chassis dimensions : $81 \mathrm{n}, \times 61 \mathrm{n}, \times 2 \mathrm{in}$. ALL COMPONENTS SUPPLIED ARE

44. 19.6 Plus 2/6 PACKING Th CARRIAGE \& INSUR. The Output Transiormer supplled is for use with a loudspeaker of 3 ohms impedance and we would suggest that the output of the cornpleted amplifer justifles the use of one of the latest W.B. H.F. Speakers which can be supplied as $\{0110 \mathrm{u}: 8 \mathrm{in}, 60 / 6$; 9in., $67 /-$; 10in., 73/6. All plus $2 / 6$ pkg., carr. ins.
To those who require thls Amplifier ready-built we can supply It at $85 / 19 / 6$, plus $3 / 6 \mathrm{pkg}$., carr., ins.

## RADID CABINET

Superb highly pollshed walnut cabinet, strongly con structed. Dimensions:-Height 171 in ; Width 17 tin Depth 9in. Shorn with dial of chassis b low.

$£ 3$. 15 . 0 Plus $7 / 6$ Carr \& Ins.


Fitted with Valve Holders, Aertal, Earth and Gram Socket;FullVision 3 Wave band Dialand Drive Assembly long, 51 in . wlde, 2 inin. deep.
42/6 Plus $1 / 6$ Pkg, Carr. \& Ins.
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## R.S.C. MAINS AND OUTPUT TRANSFORMERS

## Fully Guaranteed, Interleaved and Impregnated

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| :---: | :---: |
| Primaries $200250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. |  |
| 6.3 v. 1.5 | 5/9 |
| 6.3 v. 2 a | 7/6 |
| 6.3 v. 3 a | 9/6 |
| 0-4.6.3 v. 2 a | 7/9 |
| 12 v. 1 a | 7/11 |
| 6.3 v. 6 a. | 17/6 |
| 0-2-4-5-8.3 | 16/9 |
| 12 v. 3 a. or | 17/6 |
| CHARGED TRANSFORMERS |  |
| All with 200-230-250 v. $50 \mathrm{c} / \mathrm{s}$. Primaries: 0-8-15 v. |  |
| 1.5 a., 12/9 ; 0-9-15 v. 3 a., 16/9; 0-9-15 v. 6 a., |  |
| $22 / 9 ; 0-40-15-24$ v. 3 a., 22/9; 0-9-15-30 v.3 a., 23 . |  |
|  |  |
| TOP SHROUDED DROP THROUGH TYPE |  |
| Primaries 200-230-250 v. $50 \mathrm{c} / \mathrm{s}$. |  |
| $250-0-250$ v. 70 mA ., 6.3 v. 25 a. . ............ 12/11 |  |
| $260-6-260$ v. 70 mA., 6.3 v. 3 a., 5 v. 2 a.... 14/11 |  |
| $350-0-350$ v. 80 mA ., 6.3 v. 2 a., 5 v. 2 a.... 16/9 |  |
| $350-0-350$ v. $80 \mathrm{mAA}, 6.3$ v. 3 a., 4 v. 2.5 a.... 14/11 |  |
| $250-0-250$ v. $100 \mathrm{mA}$. ., 6.3 v 4 a., 5 v. 3 a.... 23 |  |
| $300-0-350$ v. $100 \mathrm{~mA} ., 0.3$ v.-4 v., 4 a., c.t., |  |
| $350-0-350$ v. $100 \mathrm{~mA} ., 6.3$ v. -4 v. 4 a., c.t., |  |
| 350-0-350 v. $150 \mathrm{~mA} ., 6.3$ v. 4 a., 5 v $3 \mathrm{a} . .$. |  |
| $350-0-350$ v. 150 mA ., 6.3 v. 2 a., 6.3 v. 2 a ., |  |
| E.H.T. TRANSFORMERS. 2,500 v. 5 mA , |  |
| 2-0-2 v., 1.1 a., 2-0-2 v. 1.1 a., for |  |
| VCR07, VCR517 or ACR2X .............. 35/- |  |
| 5,000 v. 5 mA .2 v. 2 a. | 39/6 |

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Primaries 200-230-250 v. $50 \mathrm{c} / \mathrm{s}$.
$250-0-250$ v. $60 \mathrm{~mA} ., 13$ v. 2 a., 5 v. 2 a. Midget type $21-3-3 i n$.
$350-0-350$ v. 70 mA ., 6.3 v. 2 a., 5 v. 2 a.... $16 / 9$
$300-0-300$ v. 60 mA .12 v. 1.5 , 5.2 a .... $18 / 9$
$250-0-250 \mathrm{v} .100 \mathrm{~mA} ., 6.3$ v. -4 v. 4 a
$0-4-5$ v. 3 a.
$250-0-250$ v. $100 \mathrm{~mA} ., 6.3$ v. 6 a., 5 v. 3 a
for R1355 conversion
$300-0-300$ v. $100 \mathrm{~mA}, 6.3$ v.-4 v. 4 a. c.t
0-4-5 v. 3 a
$350-0-350$ v. 1
$50-0-350 \mathrm{v}$ 150............................ 25/9
$350-0-350$ v. 150 mA ., 6.3 v. 4 a., 5 v. 3 a...
$350-0-350$ v. $150 \mathrm{~mA} ., 6.3$ v. 2 a., 6.3 v. 2 a.,
5 v. 3 a. ...............................................
$350-0-350$ v. $160 \mathrm{~mA} ., 6.3$ v. 6 a., 6.3 v. 3 a.,
$350-0-350$ v. $250 \mathrm{~mA} ., 8.3 \mathrm{v} .6$ a., 4 v. 8 a.,
$0-2-6$ v. 2 a., 4 v. 3 a., for Electronic Eng. Televisor

67/6
425-0-425 v. $200 \mathrm{mA},$.6.3 v. -4 v. 4 a., c.t.,
$6.3-4$ v. 4 a., c.t., $0-4-5$ v. 3 a. suitable
Williamson Amplifier, etc.
$425-0-425$ v. $250 \mathrm{~mA} ., 6.3$ v. 6 a , 6.3 v. 6 a.,
49/9
$\qquad$
EX-GOVT. E.H.T. SMOOTHING CONDENSERS $.02 \mathrm{mfd} .5,000$ v. Bakelite Tubulars .......... 1/6
$.02 \mathrm{mfd} .8,000 \mathrm{v}$. Cans
1 mfd. 2,000 v. Blocks
.25 mfd . 5,000 v. Blocks
$.5 \mathrm{mfd} .3,500 \mathrm{v}$. Cans
1 mfd . plus $1 \mathrm{mfd} 8,000$ v, large blocks
(common negative isolated)

SMOOTHING CHOKES
$250 \mathrm{~mA} ., 7-10 \mathrm{H} .200$ ohms Shrouded ...... 16/9
$250 \mathrm{~mA} ., 3 \mathrm{H} .50$ ohms.
$100 \mathrm{~mA}, 15 \mathrm{H}, 350$ ohms
80 mA ., 10 H . 350 ohms....................... $\quad$ 1/6
$80 \mathrm{~mA} ., 10 \mathrm{H} .350$ ohms ....................... $5 / 6$
$60 \mathrm{~mA} ., 10$ H. 400 ohms ........................ 4/11
50 mA ., $40 \mathrm{H} \quad 1,000$ ohms Potted
10/9

## ELIMINATOR TRANSFORMERS

Primaries $200-250$ v. $50 \mathrm{c} / \mathrm{s} .120 \mathrm{v} .40 \mathrm{~mA} . . . \mathrm{C} / 11$ 120 v. $40 \mathrm{~mA} .5-0-5$ v. 1 a.

14/9

## OUTPUT TRANSFORMERS

Midget Battery Pentode 66: 1 for 3S4, etc. $3 / 6$ Small Pentode, $5,000 \Omega$ to $3 \Omega$
Standard Pentode, $5,000 \Omega$ to $3 \Omega$
Standard Pentode, $8,000 \Omega$ to $3 \Omega$
Standard Pentode, 10,000 ohms to 3 ohms Multi-ratio $40 \mathrm{~mA} .30: 1.45: 1,60: 1$,
90 : 1, Class B Push-Pull
Push-Pull 8 Watts 6 V 6 to 3 ohms
Push-Puill 8 Watts 6V Push-Pull 10-12 Watts 6V6 to $3 \Omega$ or $15 \Omega$
Push-Pult $10-12$ Watts to match 6 V 6 to

Size approximately $12 \mathrm{in} . \times 6 \frac{1}{2} \mathrm{in} . \times 5 \mathrm{in}$. Bakelite type available in Brown or Cream. Price of Cabinets 17/6 ca
Suitable fully punched T.R.F. 3-valve and
L.T. Types

2/6 v. 1 a.h.w. $1 / 9$
F.W. Bridge Types
$\begin{array}{lll}6 / 12 \text { マ. } 1 \text { a. ...... } & 5 / 9 \\ 6 / 12 \text { v. } 2 \text { a...... } & 9 / 9\end{array}$

P.M. SPEAKERS. All 2-3 ohms. 33 in . Goodmans (Ex New Units), $10 / 9$. $6 \frac{1}{2 n}$. Goodmans, $16 / 9$.
8in. Plessey, 15/9. 8 in R.A. Heavy duty, 18 . 8in. Plessey, $15 / 9 . \quad 8 \mathrm{in}$ R.A. Heavy duty, $18^{\prime} 9$.
$10 \mathrm{in}$. Rola, $27 / 9 . \quad 10 \mathrm{in}$. Plessey, $18 / 6$. 10 in. Rola with Trans., 29/6. 12in. Truvox, 49/9.
M.E. SPEAKERS. All $2-3$ ohms, $6 \frac{1}{2} i n$. Rola field 700 ohms, $11 / 9$. 10 in . R.A. field 600 ohms, $23 / 9$. 10in R.A. field 1,500 obms, 23/9. 10in. R.A field 1,000 ohms, 23/9.

EX-GOVT. ACCUMULATORS with non-spill vents Unused and guaranteed. 2 v. 16 A.H., $5 / 9$ each, or 3 in wood carrying case $9-7-5$ in., $14 / 9$, plus 2/6 Сагт.
rectifier chassis
Suitable fully punched superhet chassis (4 valves and rect.)
Dial Scales, 2 colour, 2 waveband, station
named, glass
Dial Scales, 3 colour, 3 waveband, station named, glass
Suitable coloured Metal B $\Rightarrow$ kplates
Pointers, Double ended
T.R.F. Coils, 2 waveband with circuit

Drum Drives, complete

## THE SKY GHIEF T.R.F. REGEIVER

A design of a 4 -stage, 3 valve $200-250$ v A.C Mains receiver with selenium rectifier. For nclusion in any of cabinets illustrated above. It
consists of a variable Mu high gain H.F stage consists of a variable Mu high gain H. F stage
followed by a low distortion grid detector triode. The next stage is a further triode amplifier with tone correction by negative feedback. Finally comes the output stage consisting of a parallel connected double triode giving ample output at an extraordinary low level of distortion. Point to point wiring diagrams, instructions, and parts ist, 2/6. This receiver can be built for a maximum of $£ 4 / 16 /$ - including cabinet.

## SELENIUM RECTIFIERS

EX-GOVT. T.V. TYPE TRANSFORMERS. All $230 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$ input.
$12500-1250$ v. $250 \mathrm{~mA}, 4$ v. 3 a. $\ldots \ldots . . .25 /=$
6.3 v. 6 a., 6.3 v. 6 a., 5 v. 3 a., 5 v. 3 a., $22 / 6$
4 v. 3 a. ....................................................

400 v . C T 150 mA .4 v. 6 a., 6.3 v .6 a.,
6.3 v. $0-6$ a., 4 v. 6 a., 4 v. 3 a., 4 v. 3 a.,
4 v. 3 a., 5 v. 2 a. ................................ $22 / 9$

EX-GOVT. BLOCK PAPER CONDENSERS
$\begin{array}{llllll}4 \mathrm{mfd} . & 500 \mathrm{v} . & \ldots & 2 / 9 & 4 \mathrm{mfd} .1500 & \mathrm{v} . . . \\ 4 \mathrm{mfd} . & 450 \mathrm{v} . & \ldots & 3 / 3 & 10 \mathrm{mfd} . & 1500 \\ \mathrm{v} . & 7 / 9\end{array}$ 4 mfd .400 v . plus $2 \mathrm{mfd} .250 \mathrm{v} ., 1 / 11$.
sPECIAL OFFERS. Germanium Crystal Diodes 1/11. Midget Mains Transformers (size approx. ( $2 \frac{1}{2} \times 3 \times 2$ inin.). Screened Primary 220/240 v. $50 \mathrm{c} / \mathrm{s}$. Output, $220-0-220$ v. 40 mA . $8-3$ v. 1.5 a., 9/9.

EX-GOVT. GATHODE ISOLATING FILAMENT TRANSFORMERS. 6.3 v . to 6.3 v . c.t., $3 / 9$ ea. $2 / 9$, VS110 $1 / 9$.
i"

EX-GOVT. CATHODE RAY TUBES
VCR517 (guaranteed full picture) (carr. 5/-) 29/6 ea ACR2X (guaranteed full picture) (саr. $5 /$ ) $12 / 6 \mathrm{ea}$.
EX-GOVT. TRANSMITTER-REGEIVER TYPE TR9D, complete with all valves, only $47 / 9$, plus
carr. $5 /$.
H.T. Types. H.W. 70 v .20 mA . 90 v .20 mA . 120 v .40 mA . 250 v .50 mA. 350 v .50 mA . $250 / 350 \mathrm{v} .80 \mathrm{~mA}$. $8 / 9$

CO-AXIAL CABLE. 75 ohrms tin., 8d. yard. SPECIAL PURPOSE EX-GOVT. VALVES (GUARANTEED)
VR91, 5/9, SP61 (VR65), 3/9, VR56 3/11, $8076 / 11$, $6 \mathrm{~J} 610 / 6$, 6 SH7Met $8 / 11$, 12SC7GT 6/11, VU120A'

Push-Pu or $15 \Omega$................................
Push-Pull 15-18 Watts to match 6L6, etc., to
$3 \Omega$ or $15 \Omega$ Speaker
Push- Pull 20 Watts high-quality sectionally
wound, $6 \mathrm{~L} 6, \mathrm{KT} 66$, etc., to 3 or $15 \Omega$... $47 / 9$
Williamson type, exact to author's
specification $. . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . ~ 85 / ~$
MICROPHONE TRANSFORMERS
100:1
EX-GOVT AUTO TRANSFORMERS $50 \mathrm{c} / \mathrm{s}$ Double Wound 100 watts, $5-0-115$ 125 v . to $10-0-10-210-230 \mathrm{v}$. or revarse

18/9
15-10-5-0-215-235 v. 200 watts
Double wound $10-0-200-220-240 \mathrm{v}$.
input, 10-0-270-290-310 v. Output 200 watts

27/9
Double Wound $220 / 240$ $\mathrm{v}_{\text {. input. }}$
Output 51 v . to 250 v. 21 amps. in steps of 11 v .

## EX-GOVT. MAINS TRANSFORMERS

## 9

 $70 \mathrm{mA} .5-.10 \mathrm{H}$.$50 \mathrm{mA} 5-.10 \mathrm{H}$.
L. T. type 1 amp.

All $230 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. input 48 v .1 a . output ... $9 / 6$ Outputs $250-0-250 \mathrm{v} .40 \mathrm{~mA}$., $6.3 \mathrm{v} .2 \mathrm{a} ., 5 \mathrm{v} .2 \mathrm{a}$. $10 / 9$ $350-0-350$ v. 150 mA . 5 v. 3 a.
VALVE SCREENING CAN8. International Octal $\therefore$ piece, 10/6 doz., $1 / 3$ each.
EX-GOVT. SMOOTHING CHOKES
250 mA .10 H .50 ohms
250 mA .20 H .250 ohms. Tropicalised
250 mA .10 H .100 ohms
250 mA .3 H .50 ohms Potted
150 mA .10 H .50 ohms
100 mA .10 H .100 ohms. Tropicalised
100 mA . H 100 ohms. Tropicalised
200 mA 10 H. 100 ohms. Potted
bakelte and walnut verem cabne $90,25,30,35,50,100,120,150,180,200$, , 3 , 30 , $25,100,120,1$, $(.001 \mu \mathrm{~F}), .002 \mathrm{mfd} .(2,000 \mathrm{pfd}$.$) . All at 5 d$.

5 V 0.15 a $6 / 9 \mathrm{doz} 35 \mathrm{~V}$

VOLUME CONTROLS with long spindles all values less switch, $2 / 9$; with S.P

WIRE WOUND POTS: 30 ohms, 500 ohms 1,000 ohms, $5 \mathrm{~K}, 20 \mathrm{~K}, 50 \mathrm{~K}$ (medium lengt pindles), 2/9. 220 ohms, $2 \mathrm{~K}, 10 \mathrm{~K}, 20 \mathrm{~K}$, 50 K Preset

2in. scale, $11 / 9$.

| Tubular Types | pes |
| :---: | :---: |
| $8 \mu \mathrm{~F} 450$ v. $1 / 11$ | $16 \mu \mathrm{~F} 450 \mathrm{v}$. $2 / 9$ |
| $8 \mu$ F 500 v. $2 / 9$ | $24 \mu \mathrm{~F} 350 \mathrm{v}$. $2 / 11$ |
| $16 \mu \mathrm{~F} 350 \mathrm{v}$. $2 / 3$ | $32 \mu \mathrm{~F} 350 \mathrm{v}$. $2 / 11$ |
| $16 \mu \mathrm{~F} 450 \mathrm{v} . \quad 2 / 9$ | 32 mfd .450 v. $4 / 9$ |
| $16 \mu \mathrm{~F} 500$ v. $3 / 9$ | 64 mfd .450 v . $4 / 9$ |
| $24 \mu \mathrm{~F} 350 \mathrm{v}$. 3/3 | $8-8 \mu \mathrm{~F} 350$ v. $3 / 9$ |
| $32 \mu \mathrm{~F} 350$ v. $3 / 9$ | $8-8 \mu \mathrm{~F} 450 \mathrm{v}$. $3 / 11$ |
| 32 mfd .500 v. 5/9 | $8-8 \mathrm{mfd} .500$ v. $4 / 9$ |
| $8-16 \mu \mathrm{~F} 500$ v $4 / 11$ | $8-16 \mu \mathrm{~F} 450$ v. $2 / 11$ |
| $25 \mu \mathrm{~F} 25$ v. $1 / 3$ | $16-16 \mu \mathrm{~F} 450$ v $4 / 11$ |
| $50 \mu \mathrm{~F} 12 \mathrm{v} \quad 1 / 3$ | $16-32 \mu$ F 350 v. $4 / 9$ |
| $50 \mu \mathrm{~F} 50 \mathrm{v}$. $2 / 3$ | $16-32 \mathrm{mfd} .450 \mathrm{v} .4 / 9$ |
| Can Types | $32-32 \mu \mathrm{~F} 350 \mathrm{v}$. $\quad 4 / 9$ |
| $8 \mathrm{mfd} .450 \mathrm{v} .2 / 3$ | $32-32 \mu \mathrm{~F} 450 \mathrm{v} \quad 5 / 11$ |
| 8 mfd .500 v $2 / 11$ | 60-100 mfd. 450 v. $7 / 9$ |
| 16 mfd .350 v. 1/11 | 64-120 mfd. 350 v. 7/6 |

## MISGELLANEOUS EX-GOVT. ITEMS

Sydelock Fuses, 15 amp., 1/9. Bulgin octal type soulded Bakelite, 9 -pin or 7 -pin Plugs and resistance, $1 / 3$. pair. Earphones (Single), low
R.A

## R.S.C. 25 WATT "PUSH PULL" AMPLIFIER

Now firmly established and proving extremely popular, our Air Quality Amplifier we consider to be the best value in amplifiers offered to-day. The volume of its high fidelity reproduction is completely controllable, from the sound of a quiet intimate conversation to the full glorious volume of a great orchestra. Its sensitivity is so high that in areas of fair signal strength it can be operated straight from a crystal receiver. Entirely suitable for standard or long playing records in small homes or in large auditoriums. For electronic organ or guitar or for garden parties or dance bands.
The kit is complete to the last detail, and includes easy to follow point-to-point wiring diagrams.
Twin volume controls with twin input sockets allow SIMULTANEOUS INPUTS for BOTH MICROPHONE and GRAM, or TAPE and RADIO. SEPARATE BASS and TREBLE CONTROLS giving both LIFT and CUT. FOUR NEGATIVE FEEDBACK LOOPS with 15 db in the main loop from output transformer to voltage amplifier. Frequency response $\pm 3 \mathrm{db}$. 50-20,000 c.p.s. Hum and distortion LESS THAN o. 5 per cent. measured at Io watts. This is comparable with some of the highest priced amplifiers. Six B.V.A. valves, Marconi/Osram KT series output valves. A.C. only, 200-230-250 v. $50 \mathrm{c} / \mathrm{s}$. input. 420 v. H.T. LINE. Paper reservoir condenser. Compact chassis. Matched components

OVERALL SIZE $12 \times 10 \times 9$ in. approx. Output impedances for 3 and 15 ohms speakers.


Available in kit form at 9 Plus the amazingly low price of 9 gns. carriage 5/-

Or ready for use 50/- extra.

R.S.C. MASTER INTERCOMM. UNIT, with provision for up to 4 "Listen-Talk Back Units' individually switched. A high gain amplifien enables speech and other sounds emanating from the rooms containing remote control units to be heard at the master control. The unit is in kit formo and point-to-point wiring diagrams are supplied A walnut veneered wood or Brown Bakelite cabinet is included. Mains input is $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. H.T. line 300 v . CHASSIS IS NOT "ALIVE." Ideal - also for use as "Baby Alarm." Sound amplification 4 watts. Price only $£ 5 / 19 / 6$. "ListenTalk Back Unit'" as illustration can be supplied at 30/- each. Full descriptive leaflet 10 d .
The Master Unit can be supplied assembled and tested for 30/- extra.
R.S.C. BATTERY CHARGER KITS. For mains input $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. To charge 6 v . accumu lator at 2 amps., $25 / 9$.

To charge 6 v . or 12 v accumulator at 2 amps., $31 / 6$
To charge $6 \mathrm{\nabla}$. or 12 v . accumulator at 4 amps., 49/9.
ABOVE KITS CONSIST OF BLACK CRACKLE LOUVRED STEEI CASE, MAINS TRANS FORMER, FULL WAVE METAL RECTIFIER FUSES, FUSE-HOLDERS AND CIRCUIT. The mean charging rates are as indicated above, and complete safety is ensured by fusing of both input and output. Chargers supplied assembled and tested for $6 / 9$ extra.

A PUSH-PULL 3-4 WATT HIGH-GAIN AMPLIFIER FOR $83 / 12 / 6$, plus carr. $2 / 6$. For mains input $200-250 \mathrm{v} 50 \mathrm{c} / \mathrm{s}$. Complete kit of parts including point-to-point wiring diagrams and instructions. Amplifier can be used with any type of feeder unit or pick-up. Output is for type of
$2-3$ ohm speaker unit or pick-up. (We can supply a very suitable $2-3$ ohm speaker. (We can supply a very suitable 10 in . unit by Rola at 27/9.) The amplifier can be
supplied ready for use for $25 /-$ extra. Full supplied ready for descriptive leaflet 7 d .


GOLLARO 3-SPEED AUTOMATIC RECORD CHANGERS (brand new), type RC3521, complete with 2 plug-in Crystal P.U. heads for long playing or standard records 7, 10 or 12 in . Not intermixed. Mains input $200-250 \mathrm{v}$. Limited number available at only $89 / 15 /$-, plus carr. $5 /-$

COLLARO 3-SPEED RECORD PLAYER UNIT Type RC3/514, complete with Orthodynamic Pick-up and matching transfomer. Separate Stylii for long playing or standard records are moved into position by a switch which also makes necessary weight adjustment. Mains input 200250 v . A.C. Brand new in Makers cartons, $£ 6 / 19 / 6$, plus 5/- carr

COLLARO RECORD PLAYER UNIT. Type AC/514. Standard 10in. turntable. Speed normal 78 r.p.m. Crystal pick-up. Mains input $200-250 \mathrm{v}$ A.C. Brand new cartoned $43 / 19 / 6$, plus $5 /$ - carr

COLLARO TAPE DESK MOTORS. Shaded pole type. Clockwise or anti-clockwise. Mains input $110-200-250$ v., $31 / 6$.

# RADIO SUPPLY CO. 

R.S.C. 8-10-watt "Push-Pull " HIGH-FIDELITY AMPLIFIER A3. Complete with integral pre-amp. Tone control stage (as AII amplifier), using negative feedback, giving humproof individual bass and treble lift and cut tone control. Six Negative Feedback Loops. Completely negligible hum and distortion. Frequency response $\pm 3 \mathrm{db}$, $30-20,000 \mathrm{c} . \mathrm{p} . \mathrm{s}$. Two independently controlled inputs. Six B.V.A. valves. A.C. mains 200-230250 v . input only. Outputs for 3 or 15 ohm speakers. Kit of parts complete in every detail. £ $8 / 19 / 6$, plus $5 /$ - carriage, or ready for use, $45 /$-extra.
R.S.C. TONE CONTROL-PRE-AMP. UNIT.
complete set of parts for the construction of a very efficient but simple pre-amplifier and tone control unit. Suitable for use with any amplifer and pick-up. Fil. supply is self-contained. Size is $7 \mathrm{7}_{\mathrm{t}}-5-5 \frac{1}{2} \mathrm{in}$. approx. Full descriptive leaflet 9 d . Price, including wiring diagrams, $37 / 6$.
Or ready for use, 15 - ex ra.

## CHASSIS

18 s.w.g. undrilled alu- $16 \mathrm{~s} . \mathrm{w} . g$, aluminium, reminium amplifier type ceiver type.
 $12 \mathrm{in} . \times 9 \mathrm{in} . \times 2 \mathrm{in}, \ldots 6 / 1116 \mathrm{in} . \times 8 \mathrm{in} . \times 2 \mathrm{in} . . .7 / 6$ $14 \mathrm{in} \times 10^{2} \times 2 \mathrm{in} . .6 / 1120 \mathrm{in} . \times 8 \mathrm{in} . \times 2$ inin. $\ldots 8 / 11$ $16 \mathrm{in} . \times 10 \mathrm{in} . \times 3 \mathrm{in} . \ldots 8 / 3$
18 s.w.g. aluminium re- 16 s.w.g. aluminium, amceiver type. plifier type, 4 -sided.
$6 \mathrm{in} . \times 3 \mathrm{in} . \times 1 \frac{1}{\mathrm{in}} . \ldots .1 / 1112 \mathrm{in} . \times 8 \mathrm{in} . \times 2 \mathrm{in} . . .7 / 11$ 7 in. $\times 4$ in. $\times 2$ in $\ldots 2 / 9$ 16in. $\times \sin \times 2$ in $10 / 11$
 $11 \mathrm{in} . \times 6$ in. $\times 2 \frac{1}{2}$ in $\ldots 3 / 1114 \mathrm{in} . \times 10 \mathrm{in} . \times 3 \mathrm{in} \ldots 13 / 6$
H.T. ELIMINATOR AND TRICKLE CHARGER KIT. Consists of H.T. and L.T. transformer, H.T. and L.1. rectifiers, smoothing electrolytic choke, and steel case. For mains input of $200-250 \mathrm{v}$. Output 120 v. 40 mA . and 2 v . $\frac{1}{3}$ a. Price with circuit 29/6. OI in working order, 37/6.
PERSONAL SET BATTERY SUPERSEDER KIT. All parts for an "All Dry" Battery Eliminator. Complete with case. Supplies 90 v . 10 mA . and 1.4 v .250 mA . fully smoothed, from normal. $200-250$ v. $50 \mathrm{c} / \mathrm{s}$. mains. For 4 -valve superhet receivers. Price with circuit 35/9. Or ready for use, 42/6. Size of unit $5 \frac{1}{4}-4-1 \frac{1}{2} \mathrm{in}$.

BATTERY SET CONVERTER KITS. All parts for converting any type of battery receiver to all mains. A.C. $200-250$ v. $50 \mathrm{c} / \mathrm{s}$. Kit will supply fully smoothed H.T. of 120 v .90 v . or 80 v . at up to 40 mA , and fully smoothed L.T. of 2 v . at 0.4 a , to 1 a. Price complete with circuit and instructions only $48 / 9$. Supplied ready for use for $7 / 9$ extra.

# "MUST HAVE" BARGAINS 

RECEIVER R3II8, ideal for conversion to TV, having a built-in A.C. mains Power Pack or $180-240$ volts, is tremendously poweriul employing 7 I.F. stages of $12 \mathrm{Mc} / \mathrm{s}$ with $4 \mathrm{Mc} / \mathrm{s}$ Bandwideh and has 16 valves as follows: 6 of SP61, 4 of EA50. 2 of VR136, 1 each VRI37, P6I, 5Z4, and Y63 "MAGIC EYE" IN NEW CONDITION, only 97/6 (carriage, etc., 7/6).
"PYE" $45 \mathrm{Mc} / \mathrm{s}$ I.F. STRIP. Ready made for the London Vision Channel. Complete with 6 valves EF50 and I EASO. BRAND NEW. ONLY 70/(postage, erc., 2/6).
I.F. STRIP 194. An easily modified strip recommended for TV conserip recommended for TV constructors who want good results at
moderate cost, or for those who have moderate cost, or for those who have
built televisors but are having crouble built televisors but are having crouble
in the sound or vision receivers. Size in the sound or vision receivers. Size
$18 \mathrm{in} . \times 5 \mathrm{in} . \times 5 \mathrm{in}$., it is complete with 18 in . $\times 5 \mathrm{in} . \times 5 \mathrm{in}$., it is complete with
6 valves VR65, I of VR92, and I of VR56 or VR53. Mod. data supplied. ONRY 45/- (postage, etc., 2/6). Less valves, $19 / 6$ (post, etc., 2/6).

TELESCOPIC AERIAL, Pulls out of metal tube 15 in . long to extend to 73 in . BRAND NEW. ONLY $7 / 6$ (postage 10d.).

AMPLIFIER 208 Ideal for conversion into a high-gain TV pre-amp. Complete with 2 valves EF50. ONLY 15/(postage, etc., $1 / 6$ ).
CHOKES. $3 \mathrm{H} 40 \mathrm{~mA}, 3 / 6.5 \mathrm{H} 200$ mA., 716. 10 H 60 mA ., $4 / \mathrm{-}, 30 \mathrm{H} 100 / 150$ mA, 12/6. (Postage 1/-).

CHASSIS OF POWER UNIT 529. An ideal unit for component value or for building an amplifier, etc. Contains valveholders, resistors, potentiometer chokes, and block and tubular condensers. Housed in grey metal case, size $12 \mathrm{in}, \times 8 \frac{1}{2} \mathrm{in} . \times 7 \frac{1}{2} \mathrm{in}$. BRAND NEW. ONLY 10/- (carriage, etc., 3/6).

AMERICAN 12 v. DYNAMOTORS. Output 255 v. 60 mA . ONLY $22 / 6$. 24 v. BLOWER MOTORS. ONLY 17/6.
C.R. TUBE VCR97. Tested full screen. BRAND NEW IN MAKER'S CRATES. ONLY 42/6.

6 v. VIBRATOR UNITS. Made by the National Co. of America for use with H.RO Communications Receivers, supplying 165 v . at 85 mA ceivers, supplying 165 . at 85 mA
fully smoothed D.C. Complete with vibrator and $6 \times 5$ rectifier in black crackle cabinet size $7 \mathrm{in}, \times 7$ in. $\times 6 \mathrm{in}$. ONLY $29 / 6$ (postage, etc., $2 / 6$ ).

METAL RECTIFIERS, Selenium full wave bridge 6 or 12 volss; I amp 7/6; $2 \mathrm{amp} \mid 1 / 3 ; 3 \mathrm{amp} 12 / 6 ; 4 \mathrm{amp}$ 15/-

## CHARGER TRANSFORMERS.

Normal primaries, output $0-9.15$ volts : | amp 10/6; $2 \mathrm{amp} 16 / 6$; $3 \mathrm{amp} 18 / 6$; $4 \mathrm{amp} 20 /=$

## COMMUNICATIONS RECEIVER R. 1155

The famous ex-Bomber Command Recelver known the world over to be supreme in its class. Covers 5 wave ranges: $18.5-7.5 \mathrm{Mc} / \mathrm{s}, 7.5-3.0 \mathrm{Mc} / \mathrm{s}, 1,500-600 \mathrm{kc} / \mathrm{s}, 500-200 \mathrm{kc} / \mathrm{s}, 200-$ $75 \mathrm{kc} / \mathrm{s}$, and is easily and simply adapted for normal mains use, full derails being supplied. Aerial tested before despatch. BRAND NEW AND. UNUSED IN MAKER'S TRANSIT CASES. ONLY \&11/19/6.
USED RECEIVERS, also tested workiag before despatch, ET/19/6.
R. 1155 " N" Model. This is the latest version which covers the Trawler Band and in addition has ultra-slow motion tuning. Used, in good condition, and sested working before despatch c17/19/6.
A Facrory-made Power Pack, Output Stage and Speaker, contained in a black crackle cabinet to match the receiver, can be supplied for ONLY $65 / 10 / \mathrm{F}$. Plugs on to the receiver, and operates it immediately.
DEDUCT $10 /$ - IF PURCHASING RECEIVER AND POWER PACK TOGETHER.
Please add carriage costs of $10 / 6$ for recelver, and 5/-for Power Pack.

## INDICATOR UNIT TYPE 62A

Built on a two-deck chassis, this contains YCR97 Cathode Ray Tube with mu-meral screen, 12 valves EF50, 4 of SP6I, 3 of EA50, and 2 of EB34. An economical way of buying a VCR97 and EF50s. IN NEW CONDITION IN MAKER'S TRANSIT CASES. ONLY $99 / 6$ (carriage, etc., IO/6).

## ROTARY POWER UNITS TYPE 104

Input 12 v., Output 230 voles 65 mA , and 6.3 volts 2.5 amp ., Fully filtered and smoothed and noise suppressed. Ideal for car radio, etc. BRAND NEW. ONLY 15/- (postage, etc., 2/6).

## 100 MICROAMPS METERS

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350 v.
350 v. $-0-350 \mathrm{v} .150 \mathrm{~mA} ., 6.3 \mathrm{v} .5$ a., $5 \mathrm{v}, 3$ а., $32 / 6$.
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STANDARD S.T.c. RECTIFIERS $\begin{array}{llll}\text { RM1 } \\ \text { RM2 } & 125 V & 60 \mathrm{~m} / \mathrm{s} & 3 / 11 \text { each }\end{array}$ $\begin{array}{llll}\text { RM2 } & 125 V & 80 \mathrm{~m} / \mathrm{a} & 4 / 3 \mathrm{each} \\ \text { RM3 } & 125 \mathrm{~V} & 100 \mathrm{~m} / \mathrm{s} & 8 / \mathrm{e} \mathrm{each}\end{array}$ RM4 $250 \mathrm{~V} \quad 250 \mathrm{~m} / \mathrm{a} \quad 16 /-$ each standard cartridge fuses. $\frac{1}{2 m p .}$ 1 amp. if amp., 2 smp . and 3 amp ., 3id. each.
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#### Abstract

\section*{BAND III TEST SET}

TEST SET TYPE B.G. 3 is a wavemeter made by R.F. Equipment for the Ministry and covers $160-220 \mathrm{Mc} / \mathrm{s}$, having a $5 \frac{1}{2} \mathrm{in}$. instrument type directly calibrated dial with vernier. A $5 \mathrm{Mc} / \mathrm{s}$ crystal oscillator is incorporated for checking purposes. The instruction booklet states that the accuracy is better than $200 \mathrm{kc} / \mathrm{s}$. ( $1 / 10$ th of 1 per cent). The instrument em*. ploys 4-VR9I's (EF50) and I-VRI35 (UHF triode). It is housed in a strongly made, $12 \frac{1}{2} \mathrm{in}$. high, with hinged lid, which when opened reveals a black crackle instrument panel with controls, terminals, indicating panel with controls, terminals, indicating lamp, etc. The oscillator itself is contained in a completely screened compartment. The a completely screened compartment. The circuit diagram, giving full component values, circuit diagram, giving fuli component values, is engraved on a metal plate fixed to the base is engraved on a metai plate fixed to the base of the chassis: The instrument requires a power supply of 6.3 voles at 1.5 Amps . and 120 volts H.T. at 12 mA . which may be provided either from batteries (not provided) or from mains power pack which can be housed in the cabinet. Supplied complete with five valves, $5 \mathrm{Mc} / \mathrm{s}$ crystal, circuit diagram and instruction booklet, in tip-top condition for only $59 / 6$ plus $5 / 6$ carriage.


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CALLERS-PLEASE NOTE that we havea good selection of American and other types of communication receivers, including Hallicrafters, HRO, R.M.E., etc, in stock. All thoroughly reconditioned, re-aligned and in thoroughly reconditioned, re-aligned and in
perfect working order, which we will be perfect working order, which we
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COMMUNICATION RECEIVER RII55 for world-wide reception. Can be heard at any time during shop hours. Air tested prior to despatch. Brand new at £ $11 / 19 / 6$. A few slightly used at $87 / 19 / 6$. TRAWLER BAND. RII55N, with super slow motion drive, available at $\mathrm{C} 17 / 19 / 6$. Carriage in original transit cases $10 / 6$ extra on all models. Send $1 / 3$ for full details and circuit.
A.C. MAINS POWER PACK OUTPUT STAGE enables the RIIS5 to be used to operate speaker from 200/250 volts A.C. wirhout ANY MODIFICATIONS WHATEVER. All our Power Packs have heavy duty transformers, are complete with leads
and Jones plugs and are guaranteed for 6 months.
TYPE A in smart black metal casesize $8_{\frac{1}{4}}^{\prime}$ in. $x$ TYPE A in smart black metal casesize $8 \frac{1}{4}$ in. $X$
$4 \frac{1}{4}$ in. $x \quad 6 \frac{1}{2}$ in. Less speaker. Price $£ 4 / 10 / \%$, plus $3 / 6$ carr.
TYPE B with built in speaker in black metal case size $13 \frac{1}{2}$ in. $\times 5 \frac{1}{2} \mathrm{in} . \times 7 \frac{1}{2} \mathrm{in}$. Price $65 / 5 /$ plus $3 / 6$ carr.
TYPE C with an 8 in, R. \& A, speaker in specially designed, beautiful black crackle cabinet to match the receiver. Size $11 \frac{1}{2}$ in. $x$ lotin. $\times$ 6in. A de Luxe job. Price $66 / 10 /=$ plus $3 / 6$ carr.
U.S.A. DYNOMOTOR. 12 volts D.C input, 250 volts 60 mA . output. Weight $2 \frac{3}{4}$ lbs. Size $4 \frac{1}{2} \mathrm{in}$. $\times 3 \mathrm{in}$. diameter. Ideal for car radio, mobile amplifiers, small transmitters, etc. All tested prior to despatch. ONLY 22/6, post paid.

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THE RII32A receiver covers $100-124 \mathrm{Mc} / \mathrm{s}$ with variable tuning. Very easily altered to other frequencies. Complete with all If valyes. Requires only 250 voles and 6.3 voles when it is ready to operate. Complete circuit supplied. Only 45/-, plus $7 / 6$ carriage. BRAND NEW. Will operate from our standard Rll55 power pack using from our standard special lead, price $10 /=$ extra.

POWER PACK NO. 3. Standard 19in. rackmounted power packs for 200/250 volts mains operation. Paper smoothing, two heavy duty chokes, VU39 rectifier. Output 250 volts D.C. $100 \mathrm{~mA} ., 6.3$ volts 4 amps . Two types: Mark I with H.T. current meter at $\{4 / 4 /=$ : Mark II with H.T. current and voltmeters at $£ 4 / 10 /-$, carriage R1392, RII55, etc. Lead for any specified set with Jones plugs, $10 /=$ extra. All power packs guaranJones plugs, $10 /-$ extra.
teed in working order.

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$29 / 6$ plus $2 / 6$ post. Filament Transformers$29 / 6$ plus $2 / 6$ post. Filament Transformers-1" 12 volts $1 \frac{1}{2}$ Amps., 6.3 volts $1 \frac{1}{2}$ Amps. Type 12 voles $1 \frac{1}{2}$ Amps., 6.3 volts $1 \frac{1}{2}$ Amps. Type rype $7 / 6$ each.
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| Goodmans Axiom 101. 8in. L/Spkr. | ¢6 12 | I | ¢2 4 | 1 | 15 |  | 8 |  |
| Garland Amplifier ACIIA | ¢6 12 | 6 | $\underline{¢ 24}$ |  |  |  | 8 |  |
| Collaro AC47. | 6613 | 4 | ¢2 4 | 6 |  |  | 8 | 2 |
| Wharfedale Golden CBS. IOin. Speaker | 68. | 7 | E2 15 | 7 | 19.11 |  | 10 | 2 |
| Goodmans Audiom 60 Speaker | ¢8 12 | 6 | ¢2 17 | 6 | ¢10 |  | 10 | 7 |
| B.S.R. 3 -speed single player. GU4 | E9 5 | 0 |  | 8 | ¢ 12 |  | 11 | 4 |
| Connoisseur Pick Up, 2 heads |  | 6 | 531 | 10 | El 2 |  | 11 | 4 |
| Leak Pre-amplifier | 69.9 | 0 |  | 0 | 412 |  | 11 | 7 |
| Wharfedale WI2/cs. 12 in . | ¢9 15 | 0 | E3 5 |  | \& 3 |  | 11 |  |
| Goodmans Axiom 150, Mk. II | £10 5 | 6 |  | ${ }^{6}$ | ¢ 47 |  | 12 |  |
| Leak " Varislope " Pre-amplifie | 61212 | 0 | 44 | 0 | ¢1 $10 \sim$ |  | 15 | - |
| Garrard 3-speed Auto-changer. R.C.75A | ¢15 8 | , | 652 | 8 | ¢ 1610 |  | 18 | 10 |
| Garland Amplifier, AC IV | ¢15 15 | 0 | $\pm 5 \quad 5$ | 0 | ¢1 17 |  | 19 |  |
| B.S.R. " Monarch ' 3 -speed Auto-change | ¢16 10 | 3 | 6510 | 1 | $\mathrm{El}_{1} 196$ |  |  | 1 |
| Garrard 3-speed Auto-change. R.C. 80 | ¢17 1 | 3 | 6513 | - | $¢ 2010$ |  | 0 |  |
| Lane Tape Desk, Mk. IV | 61710 | 0 | 4516 | 8 | 12 |  | 1 |  |
| "Unitelex" Record/Playback Amplifier. PR3A | ¢20 0 |  |  |  |  |  |  |  |
| Connoisseur 3-speed unit | ¢21 17 | 3 | 67.5 |  | 62123 | E1 | 6 |  |
| Truvox Tape Deck, Mk, | ¢23 2 | 0 | 6714 | 0 | 6215 |  |  |  |
| Leak "P Point One" Amplifier | ¢28 7 | 0 | 69. |  | 637 |  |  |  |
| Q.U.A.D. Amplifier with pre-amplifier... | ¢35 | 0 | ¢11 13 | 4 | $4^{4}{ }^{3}$ |  |  |  |
| Herald Tape Recorder | 650 |  | 41616 |  | 6517 |  | 1 |  |
| Unitelex "Prima" Tape Recorder. (Described above) | 65917 | 0 | 61919 | 0 | 673 |  |  |  |

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EF50, 6/- : ENGLISH OCTAL, $3 \%$ per doz. SCREEN CANS for B9A, B7G, 6/ doz: PAXOLIN-B7G MAZDA 4-pin UX ............................................................. BRAND NEN 5-pin Chass BULGIN. P.73, Plug and Socket, $2 / 9$ each; P74, Plug and Socket, 2/6: P200, Plug and Socket, 2/-: Rotary Switches S.255. 2/- ; Dolly Switches, S.267, 2/-: Dólly Switches, S.259, $1 / 6$; Standard Switches, Ex-Govt., On-off
POST OFFICE LAMP JACKS, No. 10
Lamp Covers for same.
F. CHOKES $300 \mathrm{w}, 60 \mathrm{~mA}$. CH5

OUTPUT TRANSFORMERS. Multi Ratio, $5 / \%$; Pentode
VALVE SCREEN CANS for Standard Valves
DRUM DRIVES. $4 \frac{1}{i n}$
DRUM DRIVES. $4 \frac{1}{2}$ in. ….................................
WESTECTORS. WX6, WX12, WI, W 12, W4
SIGNAL LAMP HOLDERS P/M, complete with adjusting
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AIR SPACE TRiMMERS. Preser and spindle types, 5PL, 10PF, 15PF, 20PF, 25PF, 50PF, 75PF, 15/ ; 100PF Preset
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$8-p i n, 3 / 6: 10-p i n, 4 /-$
NUTS. $8 \mathrm{BA}, 3 / 6 ; 6 \mathrm{BA}, 2 / 6 ; 4 \mathrm{BA}, 3 /=; 2 \mathrm{~mA}$
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2/6 each

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\begin{array}{cc}
2 /- & \\
3 /= & " \\
2 / 6 & \cdots \\
1 / 9 & " \\
5 / 6 & \\
15 / 6 \\
2 /- & \text { doz. }
\end{array}
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4/-"
2/6"
$4 / 6 \quad$,
$2 / 6=$
$9 / 6$
2/6 eäch
$12 /=$ doz.

3/-
1/6 pair

1/6 each
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1/- each 9/- doz. 15/-
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18/-
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AR88LF, ARB6D, CRI00, from stock. RII55 RECEIVEnS, new. A.C./D.C. MOTORS, suitable for sewing machines, 4716 each. A.C./D.C. 12 v.-15 v. MOTORS, long spindle for models, $15 /$ e each. 20 WATT P.A. RACK MOUNTING AMPLIFIERS, complete with power pack, 200/250 v. A.C., less valves, £6/10/-. Valves2 type PX25, I MH4 and I MU14, E2/15/-per set.
NEW M/C'MICROPHONES, hand type, with 12 yds . heavy duty screened cable, $£ 3 / 15 /$ - each,
B.C. 221 FREQUENCY METER, from stock Many items of American equipment available.
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 ach.
COSSOR DOUBLE BEAM OSCILLOSCOPE, perfect, 633 , G.E.C. 7 WATT V.H.F, MOBILE TX/RX. Complete with 12 v. rotary p/pack, $80.9,81.1$ and $81.3 \mathrm{Mc} / \mathrm{s}$, special offer, $\mathbf{E 3 0}$. EDDYSTONE 640 RECEIVER. Perfect, at $£ 22$.
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LARGE STOCKS OF MOTORS. A.C./D.C. and A.C., $1 / 16$, 1/12, $\frac{1}{2}, \frac{1}{2} \mathrm{~h} . \mathrm{p}$
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RECTIFIER UNITS. Westinghouse metal rectifiers, transformers. Output 50 volt $\frac{3}{4}$ amp. 50 volts $1 \frac{1}{\frac{1}{2}}$ amps., 50 volts $3 \frac{1}{2}$ transformers. Output 50 volt $\frac{4}{4}$ amp. 50 voits $1 \frac{1}{2}$ amps., 50 volts $3 \frac{1}{2}$
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INVERTBRS. Known as Motor Generator type 7. Inpat 22/29 v. D.C. Output voltage regulator cyd filter unit mounted beside generator on the same barbon pile voluge regulator and fiter unit mounted beaide generator on the same base plate
Overall dimensions approx, $9 \times 8 \times 51 \mathrm{~m}$, and weighe only $15 t 1 \mathrm{~b}$. In new and unused condition. PRIOE $85 / 10 /$.
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CABINET SPEAKERS. 8in. moving coil P.M. speaker fitted in wooden cabinetp less back, $18 \mathrm{in}, \mathrm{sq}, \times 8 \mathrm{in}$. deep with sraall carrying handle at top. In used condition but speaker tested to ensure in good working order. BARGAIN PRICE. Only 17/6, carriage 3/6
SCREENED WIRE. Copper conductor $1 / 048$, permanoid insulated. BargaLoclearance
12 yards, $5 / 8 ; 25$ yards, $10 / 8 ; 50$ yards, $17 / 8 ; 100$ yards, 25/3 GOLDHAWK ROAD, (Dept. M.W.)
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## 

 5 volt 2 amp., $82 / 6$.Drop thro' $350-0-350$ ₹. $70 \mathrm{~mA}, 6 \mathrm{~F}$ 2.5 amp ., 5 v. 2 amp., $14 / 6$.

Drop thro' $250 \cdot 0 \cdot 250$ v. $80 \mathrm{~mA} ., 6$ च. 3 amp., 6 จ. 2 amp., $14 / 6$.
Drop thro' $110-11060 \mathrm{ma}$., © F .0 .6

## amp., 8/6.

$280-0-280$, drop through, 80 mA .
6 v .3 amp. 5 F .2 amp. 1416. 6 v. 3 amp., 5 จ. 2 amp., $14 / 6$.
$250-0-25080 \mathrm{~mA} ., 6 \mathrm{v} .4 \mathrm{mp} ., 14 / \mathrm{-}$ Pri. 230 . Sec. 200-0-200 35 mA . 6 F. 1 amp., 8/6.
Pri. $200 / 250$ F., secondary $3,4,5,6$,
$8,9,10,12,15,18,20,24$ and 30 volt at 2 ampio. 131 -.
Drop thro' $280-0-280,200 \mathrm{ma}, 6 \mathrm{v}$. 5 ampi., $\delta$ v. 3 amps., $27 / 6$.
Drop thro' $270-0-27080 \mathrm{~mA}, 6 \mathrm{v}$. Drop thro' $270-0 \cdot 27060 \mathrm{~mA} ., 6 \mathrm{v}$. 3 amp., 11/6.
Heater Transformer. Pri. 230-250 $v$. 6 v. $1 \frac{1}{2}$ amp., $8 /-; 2$ v. $2 \frac{1}{2}$ atnps., $5 /-$ lnsulated, $8 / 6$. P. \& P. each $1 /$-.
Input $200 / 250$ sec., $\nabla, 6.3,6$ arop., 4 V. $1000-0-1000$ T. 250 mA iv. 3 amp 37/6. P. \& P. $5 /$.
P.M. SPEAKERS (closed fleld) leas

## 211 n. 31 n. 51 n.


P. \& P. on the above $1 /$ - each.
10in. leas trans., 1916. P. \& P. $1 / 6$

Truvor BX11. 12in. P.M. 3 ohm speech coil, 45/- P. \& P. $3 / 6$.
${ }^{6+i n}$, M.E. Speaker, 1,000 ohm field R. \& A. T.V. Energised 6ifn. speaker coll 175 ohms. Requlres a minimum 150 mA . to energise, maximum current 250 mA . $15 \%$ P. © P. $2 /$
Extension Speaiker Cabinet, in contrasting wainut vencer, vize $15 \times 10 \mathrm{ln}$. Will take 6 or 8in. apeaker $17 / 6$
P. \& P. $2 /$.
Completely built All-dry Malns Unit by famous manufacturer, $200 / 250$ v. Metal casesize $8 \times 5 \times 3 \mathrm{in}$, incorporating Westinghous
med., $16 \times 24$
metal.
mectiflers, 3
mains trang., mid., $16 \times 24$ mfd., mains trans., ${ }^{3}$ mA ., $1.4 \mathrm{~V}_{\text {. }} .25 \mathrm{amp.g} 39 / 6$. P . \& P 2/6,
Volume Controls, Long spindite less switch, $50 \mathrm{~K}, 500 \mathrm{~K}, 1 \mathrm{meg} ., 2 / 6$ each. P. \& P. 3d, each.

Volume Controls. Long spindle and witch $1,1,1$ and 2 meg $4 /-$ each: 10 K . a 60 K ., $3 / 6$ each. and 1 mek. long splndle double pole owitch, minia ture, $5 /=$ P. \& P. 3d. each.
Trimmers, $5-40$ pf., 5d.; 10-110, 10-250. $10-450 \mathrm{pf}_{4}, 10 \mathrm{~d}$.
 Line Cord, 2 -way 0.3 amp., 60 ohms perfoot. 1/3 per yard.
Twin-Gang .0005 with feet, size $3 \div \times 3 \times 1$ inn., 66.
3-gang .0005 , with feet, aize $4 \times 3 \times 1110$
$7 / 6$. $7 / 6$.
Hoover Fariable Speed 600-1,200 revs. Tape Recording Motor. Silent runaing, $200 / 250$ F. A.C. Shaded pole
 PERSONAL SEOPPERS ONLY. 9in. Enlarger 1776, 12in., 27/6.
Germanium Crystal Diode, $2 / 3$ post Germanium Crystal Diode, $2 / 3$ post
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T.V. Width Controls, 3/6.
T.V. Sub Assembiy, all-chassic, 12 n. $x$ 31 ln . with frame osc., line oac., 12 mfd
275 wkg. Metrosil, 8 condensers. resistoreand tag penel $15 /=$, p. \& p. 1/6. CRYSTAL PICK-UP by famous manufacturer complete with eapphire trailer needle aud vol
Amplifer case, black rexine covered, leather carrying handle, chrome plated corners, rubber feet, feltinned, detachable 11 d . Erternal dimensions, $131 \times$ $13 \neq 9$ in., \&1. P. \& P. 2/6. Mains Droppers, 0.3 smp, 460 ohms. tapped 280 and $410,1 / 6 ; 0.2$ amp., $\begin{array}{ll}717 & \text { ohms. tapped } \\ \text { vitreous, } 1 / 6: 0.9 \text { at } 100 \text { ohms, } \\ 950\end{array}$
 0.2 ang. 1,000 ohma, vitreous, tapped,
$2 / 6$; Vitreose .3 mpp. 700 tapped 2/6; Vitreous , 3 1mp. 700 tapped

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CABINET, asillustrated, $114 \times 6 i \times 51$ in. in wainut or cream complate with TRF new waveband, backplate, drum pointer, spring, drive oplndle, 3 knobs and beck,

AS ABOVE, with superhet chaseis, $23 / 6$. P. ${ }^{\text {\& }}$ ABOVE.

AS ABOVE, complete with new speaker to fit, and O.P.tran $8,35 /$. P. . . P. P.
superhet chassis, $36 / 6$, with
 gang with trimmer,
coill, $5 / /$ : 3 obsolete ex-Govt. valves, 3 vh and circuit, $6 / 6$; heater trane., $8 /-$; volume control with, wistch, 3/6; wavechange switch, $2 /-; 32 \times 32 \mathrm{mfl} ., 4 /=$; bias condenser, $1 /-:$ resistor $\mathrm{kit}, 2 /-$; condenser kit, 4/-
 control with switch, $4 /$-; wave-change switch, 2/6; hcater tran. 1 , $7 / 6 ; 4 \mathrm{~V} / \mathrm{h}, 1 / 6 ; 4$ obsolete ex-Govt. val ves, metal rectither a-d X tal diode with olrcult, $14 / 8 ; 25 \times 25$ mid., $1 /=; 16 \times 16$ mfd., 3/3; condeneer kit (17), 7/6; resistor hit (14), 3/6.
USEd 4-Valye plus metal rec. A.c. Mains $230 / 250$ SUPEREET. Falve line-up: $6 \mathrm{~K} 8,6 \mathrm{K7}, 607$ and 6 F 6 . Medium wave, in mahogany cabinet, size $14 \xi \times 9 \times 7$ in. These have
been checked and are, in first-clasa working order, and Arst-class performance. 61 in . P.M. been checked and are in first-
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FULLY SHROUDED MAINS TRANSFORMER, Input 110,250, 8e8。350-0-350 175 millis 6.3 v . $7 \mathrm{mmp} ., 5$ マ. 3 amp. $35 /$. P. \& P. $3 /=$.
FULLY SHROUDED PUSH-PULL TRANS. Pri. 6,000 ohme, se .16 ohme (2 L6e*ela puehopull) £1. P. \& P. $2 /$-.
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FULLY SEROUDED CEOEEE F Henry 120 mills, $8 / 6$. P. \& P. $2 /$-.
These last four item by very famous manufacturer.
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COMPLETELY BULLT STONAL GENERATOR. Coverage $110 \mathrm{Kc} / \mathrm{s} .-320 \mathrm{Kc} / \mathrm{s}$, $300 \mathrm{Kc} / \mathrm{s}$.$900 \mathrm{Kc} / \mathrm{s} ., 900 \mathrm{Kc} / \mathrm{s} .-2.75 \mathrm{Mc} / \mathrm{se}, 2.75 \mathrm{Mc} / \mathrm{s} .-$ $8.5 \mathrm{Mc} / \mathrm{s}, 8.5 \mathrm{Mc} / \mathrm{a}-25 \mathrm{Mc} / \mathrm{s} ., 07 \mathrm{Mc} / \mathrm{s} .50 \mathrm{Mc} / \mathrm{B}$,

 400 c.p.s. to a depth of 30 per cent., modulated or unmodulated R.F. output contlinuously variable 100 militrolti. Black crackle finíshed case and white tront panel. P. * P. 4/-.
24/5/-, or $34 /-$ deposit and. e4/5/-, or 34/- deposit and. 3 monthly
paymenta of 21 . buenta of $£ 1$.

CONSTRUCTOR'S PARCEL NO 1
 and plated 18 gauge, Vha, IF plate, 2 eupporting brackets, 3 waveband scales, new wavelength station names. Bize ofscale $11 \ddagger \times 4$ in., drive aplndle drum, 2 pulleys, pointer, 2 buid ctal valve holders, 4 knobs and pair of $465 \mathrm{IFs}, 16 / 6$. P. *P. $1 / 9$.

AS ABOVE, but complete with $16+16$ mid. 350 wkg. and seml-shrouded drop thro' $250-$ 200-250, and twinogang, 31/6.


CONSTRUCTOR'S PAECEL. As No. 1 , plus $16 \times 16 \mathrm{mfd} .350$ wkg., semi-shrouded drop-thro 50.0-250 $60 \mathrm{~mA} ., 6.3$ V., 3 a., 5 .., 2 a., twin gang and 6 L.M.S. superhet colle complete with trimmers and trackiag condensers with clrcuit. $82 / 5 /=$, plus $3 / 6$ postand pkg.
BATTERY CEARGER KIT comprising metal case $4 \frac{1}{4} \times 5 \times 4 \frac{1}{f} \mathrm{In}_{\text {, tranaformer } 230 / 250} \mathrm{~T}$., and metal rectlter. Will charge 6 or 12 v. battery $1 \ddagger$ amp. $19 / 6$. P. \& P. $2 / 6$.

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$300-0-30060 \mathrm{~mA} .6 .3$ ₹. 1 a., tapped at v. 6 $350-0-35075 \mathrm{ma}$
6.3 F. $1 \mathrm{~m}, 13 / 6$.
$350-0-35070 \mathrm{~mA} .47 .5 \mathrm{~s} 4 \mathrm{~F} .2 .5 \mathrm{~A}$. C.T.4. 18/6. P. \&. P. on the above $500-0-500$
6.3 ₹. О.T. 2 д. 5 v. С.T. 2 д., $27 / 6$.
$500-0-500125 \mathrm{~mA} .4$ т. C.T. 4 a. 4
C.T. 4 ๓. 4 ซ. C.T. 2.5 a., $27 / 6$.
$500-0-500250 \mathrm{~mA} .4$ ซ. O.T. 5 а. 4 v. O.T. Б \&. 4 F. G.T. 4 a., $39 / 6$
P. P. on the above transformers 8/-.

Ling and E.E.T. translormer 0KVA, using ferrocart core complete with oult-inlline and width control. Mounted
 P. \& P. 2/6.

Valve Holders, moulded octal Mazda. and loctal, 7d. each. Paxolin, octal Mazds and Ioctal, 4d. each. Moulded B7G, B8A, and B9A, 7d. each. B7G moulded with screening can, 1/6 each $16 \times 24350 \mathrm{wkg}$.
4 mfd., 200 wikg.
40 mid.
$16 \times 8 \mathrm{mfd}, 500 \mathrm{wkg}$.
$16 \times 16 \mathrm{mfd} ., 500 \mathrm{wkg}$.
$16 \times 16 \mathrm{mfd} ., 500 \mathrm{wkg}$.
$16 \times 16 \mathrm{mid}$. 450 wkg .
$32 \times 32$ mid., 350 wkg. 350 wikg. and 25
mid. 25 wkg .
$250 \mathrm{mid} ., 12 \mathrm{Fkg} . \mathrm{wkg}$
16 mid. 500 wkg., wire ends
8 mid., 500 v. wkg., wire ends
8 mfd., 350 T. wkg., tag ends.
100 midd .350 wkg.
$100+200 \mathrm{mfd} ., 350$ wkg.
$16+16 \mathrm{mfd} .350 \mathrm{wkg}$.
Ex-Govt. 8 mid.. 500 v. wkg.
size $34 \times 1$, 2 for
$16 \times 32$ mid., 250 whg.
$16 \times 32$ mid., 250 wh
$50 \mathrm{mfd} ., 180 \mathrm{wkg}$.
$85 \mathrm{mfd} ., 150 \mathrm{wkg}$.
$60+100$ mifd., 230 wkg.
50 mfd . 12 wkg. $32+32$ mid., min. 275 wk
50 mid., 50 wk g
Minlatare wire ende moulded
100 pf., 500 pf., and 001 ea.
Combined 12in. masls and escutoheon in lighely tinted perspex. New arpect edged in brown, Fits
cabinet, $17 / 6 . \mathrm{P}$ \& $\mathrm{P} .2 /$ Frame Oscillator Blooking Trans. $\quad 4 / 6$.
Trube Mounting Bracket, size $91 \times 4 i / n$. 12in. tube clamps, 2/m.
8 moothing Choke, 2 henry 150 mA , $3 / 6.250 \mathrm{~mA}$. S henry, $5 /-9250 \mathrm{~mA}$.
10 henry, $10 / 6$; B henry 250 mA ., 60 10 heary,
P.M. Focus Unit for any 9 or 12in. tube except Mazda 12in., with vernie adjuatment, $15 /$ - P. \& P. 1/6.
P.M. Foous Unit for Mazda, $121 \mathrm{n}_{\text {, }}$, with Vide Angle P.M. Focus Units F. $1 / 6$ adju Angle P.M. Focus Units, Vernie Energised Focus Coil. low reaistance mountling bracket, $17 / 6$ plua 2 - P. P. \& $\mathbf{P}$ Scan Coils, low line 10 w impedance irame, complete with O.P. transformer.
17/6. P. \& P. 2/-.

Ion Traps for Mullard or Englith Electric tubes, $5 /$ post pald.
465 ke. I.F.s, size $27 \times 1$ in. $Q .110$ removed from Amertcan equipment 5/- per pair. Standard 465 Ko. iron-
 $465 \mathrm{Kc} . \mathrm{IF}_{\mathrm{a}} 3 \mathrm{3} \times 11 \times 1 \mathrm{lin}$, per pr $9 / 6$.
Lson-cored 465 Ko. Whistle Qlter, 2/6. OUTPUT TRANSFORMERS. Shandard type $\mathrm{E}, 000$ ohms $\mathrm{lmp} ., 4 / 8$. $42-1$ with ture 42-1, 3/3. Multiratio 3,500 7,000 and 14,000 , 5/6. 10 -watt push pull, oV6 matching 7/o. $90-13 \mathrm{chm}$ speech coil, 6/6.
PUBR-BACK CONNECTING WIRR Doz. Yds 1/6, post paild.

STANDARD WAVE-OEANGE SWITCEES, 4 -pole 3 -way, 1/9; 5 -pole 3 -way, 1/9; 3-pole 3 -way, 109; 9-pole 3-way, 3/8: Mindature type, long spiudle 2-pole 16 each $P$ \& P. 3d

485 KC. MIDGET L.F.s. Q.120, ,ize 1 Hn.long, In. Wide, Ifin deep by very famoze manatactarer. Pre-aligned 18/

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If you have outstanding ability and would like to share in the unique part this company is playing in aviation progress, our Personnel Department will be pleased to know of your qualifications. Immediate Vacancies are limited, but others occur from time to time due to continued expansion. Only men of high calibre, experience and technical attainment will meet the requirements.

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Minimum requirements are higher national certificate, but an engineering degree is preferred.

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Should. have experience in electronic servo systems and, preferably, experience of magnetic amplifiers.

WRITE, AS FULLY AS POSSIBLE, TO THE DIRECTOR OF ENGINEERING, GROUP D, ROTAX LIMITED, CHANDOS ROAD, WILLESDEN JUNCTION, LONDON.

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[^0]:    - Second Report of the Television Advisory Committee. H.M.S.O Is

[^1]:    * Pye Telecommunicarions, L.d

[^2]:    * An RC Circuit Giving Over-Un'ty Gain", C.L. Longmi, e. TeleTech, April 1947

[^3]:    * Grundig (Gt. Britain) Lid.

[^4]:    ' Elect. Engng., N. Y. March 1949

[^5]:    " Transistor Research Bulletin," a new publication cover'ng progress in the field of eransistors, crystal diodes and other semi-conductor devices. The December, 1953, issuc contains information about developments in Germany and a semiconductor bibliography. From National Scientific Laboratories, 2010. Massachusetts Avenuc. N.W'. Washington, 6, D.C., U.S.A.

[^6]:    * The Plessey Company.
    $\dagger$ "Frequency Shift Keying "; Wireless World, November, 1948

[^7]:    *Research Laboratories, General Electric Company.
    $\dagger$ Royal Aircraft Establishment.
    $\ddagger$ Salford Electrical Instruments.

[^8]:    * General Electric Company (Research Laboratories).
    "Wireless World," March 1953. p. 131.

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    for Car Badio or Electric Razors, 22/6.

[^16]:    PERSONAL PORTABLE CABINET. In cream-coloured plastic: aizc $7 \times 4 \neq 31 \mathrm{~m}$. Comp post and pkg. 1/6.
    31n. P.M. Speaker to fit above, $10 / \mathrm{m}$, Mindature output transformer, $5 / \mathrm{m}$ Mimature wavechange awitch, $1 / 6$. Minlature 1 -pole 4 -way used as Volume and Off, 1/6. 4B7G valve holders, 2/4. Midget twin gang in. dia., in. long and pair medium and iongwave 8/6. Condenser Kit, comprising 11 minlature condensers, 3/6. Resistor Kit comprising 16 miniature resistors $4 /=$. The above recelver (leas valve and batteried could be built for approximately $51 /=$ All valves to ault above sralleble. Poln to Point Wiring Diar gram $1 /$

