## Wireless World

 RADIO AND ELECTRONICS

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# Wirtless World 

Managing Editor: HUGH S. POCOCK M.I.E.E.
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PUBLISHED MONTHLY: Price 2/- (last Thursday of preceding month) by ILIFFE \& SONS LTD., Dorset House, Stamford St., London, S.E.I. Telephone: Waterloo 3333 (60 lines). Telegrams: "Ethaworld, Sedist, London."

ANNUAL SUBSCRIPTION:
Home and Overseas, fl 6s. Od.; U.S.A. and Canada, $\$ 4.50$

BRANCH OFFICES:
Birmingham:
King Edward House, New Street, 2.
Coventry: 8-10, Corporation Street.
Glasgow: 268, Renfield Street, C. 2.
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The line output stage of a television receiver imposes conditions which call for a valve having properties which differ considerably from those of a normal output valve.

The essential requirements for a line output valve are, first, that it shall be capable of supplying a large peak anode current when operated at a low anode voltage, and second, that the insulation of the anode shall be capable of safely withstanding the high voltage peaks, which may amount to 5 KV to 7 KV , occurring during the fly-back period:

Satisfaction of the first of these requirements is mainly a matter of the geometry of the valve; the second can be assured only by careful mechanical design.

The PL81 is an all-glass power pentode on the B9A (Noval) base, specially designed for use as a line output valve. Its heater is rated at $0.3 \mathrm{~A}, 21.5 \mathrm{~V}$, and is therefore suitable for series operation in transformerless receivers intended for A.C. or D.C. mains supply.

It will deliver a peak anode current of 350 mA at an anode voltage of only 70 V and a screen voltage of 170 V . The efficiency of the yalve is thus intrinsically high. By adopting the booster diode circuit whereby the H.T. line voltage is increased, the required output can be obtained at a still lower anode voltage, so that the already high efficiency of the stage is further improved.
The necessary high insulation resistance to withstand a peak voltage of 7 KV has been obtained by special internal construction, and by bringing out the anode connection to a top cap so that the use of a specially insulated valve-holder is not necessary. Even so, the seated height of the valve is only 75 mm and the overall diameter 22.2 mm -dimensions only slightly larger than those of the voltage amplifying pentode in the same range. The production of a valve of this output and high insulation in so small an envelope may be counted as something of an achievement.

## RATINGS \& CHARACTERISTICS

| Heater Suitable for Series operation, A.C. or D.C. |  |
| :---: | :---: |
| $V_{h}$ | 21.5 V |
| In | 0.3A |
| Capacitances |  |
| $c_{\text {in }}$ | 14.3 ${ }^{\mu} \mu \mathrm{F}$ |
| cout | $6.5 \mu \mu \mathrm{~F}$ |
| $\mathrm{C}_{\mathrm{a}-\mathrm{gl}}$ | $<0.4 \mu \mu \mathrm{~F}$ |
| $\mathrm{c}_{\mathrm{g}} \mathrm{l}$-h | $<0.2 \mu \mu \mathrm{~F}$ |
| Characteristics |  |
| $V_{\text {a }}$ | 170 V |
| $V_{83}$ | OV |
| $V_{\mathrm{g} 2}$ | I70V |
| $\mathrm{V}_{\mathrm{g}}$ | -22V |
| $\mathrm{Ia}_{1}$ | 45 mA |
| 1 g 2 | 3 mA |
| gm | $6.5 \mathrm{~mA} / \mathrm{V}$ |
| $\mu_{\mathrm{g}}^{1-\mathrm{g} 2}$ | 6 |

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MVM 149

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VOL. LVI. No. 11.
NOVEMBER 1950

## Valve Symbols

BRITISH Standard 1409:1947" Letter Symbols for Electronic Valves," issued some three years ago, was stated to be " based on proposals prepared with a view to standardizing the letter symbols used by valve manufacturers in their catalogues and other technical literature." At the time, Wireless World-not unreasonably, we hope-took the publication of this specification as implying a desire that the symbols proposed should be generally adopted by a wider circle. So, in spite of typographical difficulties, we accordingly used the symbols as far as was practicable.
Having gone thus far in the interests of uniformity, it comes as something of a shock to us to see the revised specification B.S. 1409:1950, just issued.* This publication, under the same title as its predecessor, carries on its cover the prominent statement " Mainly for use in valve catalogues and similar technical literature." That seems to put forward a highly undesirable principle: are we to have a state of affairs where users of valves (which means all of us) are to learn two languages-one for use in dealing with manufacturers and another in reading (or writing) technical journals?

## Stereophony

OUR correspondence columns bear witness to the widespread interest that exists in stereophonic broadcasting. Stereophony will no doubt attract still more attention next year, as we believe the multi-channel principle is to be used in the sound-reproducing equipment to be installed at the Festival of Britain. That being so, it is a pity that a plea for B.B.C. experimental transmissions on a binaural system, made in this journal last March, was so uncompromisingly turned down. The B.B.C. Chief Engineer, in a letter published in our April issue, pointed out the admitted complications of the system, and expressed doubts as to

[^0]whether the Corporation would be justified in spending money on a specialized service of interest to a relatively small number of listeners. The French broadcasting authorities were, however, less pessimistic about stereophony, and staged an experimental transmission a few months later.
We are certain that interest in this subject is greater than the B.B.C. imagines, and venture to suggest the Corporation should make another attempt to find ways and means of staging experiments in which listeners can participate, as they did in France.

## Too Many Hosses?

THOUGH the uproar provoked by the withdrawal of the political play "Party Manners" from the B.B.C. television programme was a regrettable and unworthy episode that by now has been largely forgotten, it raised a matter of principle that will, we hope, be borne in mind when the B.B.C. Charter comes up for renewal next year. Wiveless World does not concern itself with details of programme organization, but we deplore anything that tends to reduce the status of British broadcasting.

As we see it, the B.B.C. attitude over this matter betrayed that weakness against which the Corporation should be most strongly on its guard-timidity This point of view was admirably expressed in a leading article in The Times of roth November "The listening public will always be behind them [the B.B.C.] in refusing to turn broadcasting into a colourless bore for the sake of this or that cran': or fanatic. There would be no intellectual or aesthetic worth in broadcasting if it were controlled to please every shade of minority opinion. Censorship by timidity can be as effective as the blue pencil of a dictator." As things are, the organization of the B.B.C. tends to favour the timorous. Heads of "Programmes" should have sweeping powers, and be discouraged from passing on the responsibility for awkward decisions to higher authority.

# Sidelights on <br> Loudspeaker Cabinet Design 



Experimental two-section vented cabinet with top half tilted back to show sound-absorbing material covering the communicating window.

IT is not the purpose of the present article to enter deeply into the mathematical treatment of the problems of loudspeaker cabinet design, since this field has already been covered in standard works on applied acoustics as well as in a number of articles in the technical press. Instead, it is proposed to discuss certain aspects of the subject which do not always receive full attention in the literature, and to illustrate these by a few experimental results obtained in the course of development work on cabinets during the period 1938-1947.

## Standing Wave Effects

The subject of loudspeaker cabinet design is usually approached by a simplified theory, in which the mechanical properties of the cone unit and the acoustic properties of the enclosure are represented by a set of inductances, capacitances and resistances. Equivalent circuits containing these elements can then be drawn and the basic features of the design calculated.

Clearly, this simplified theory is only valid at frequencies so low that the wavelength of the sound is large compared with the dimensions of the cabinet. At higher frequencies, at which this condition is not fulfilled, the cabinet volume does not behave as a

## Acoustical and Electrical Damping in

Closed-cabinet Loiudspeaters

By D. E. L. SHORTER,* B.Sc.(Eng.), A.M.I.E.E.

lumped acoustic capacity, and standing wave effects may appear unless the enclosure is adequately damped by sound absorbent material.

Unfortunately, the performance of a cabinet at these higher frequencies cannot be readily calculated, and in the absence of facilities for response measurement the degree of damping actually achieved is often in doubt.

The full extent of the standing wave effects in loudspeaker cabinets is not always appreciated. Cabinets intended to give good response down to $50 \mathrm{c} / \mathrm{s}$ generally have at least one dimension which is equal to a half wavelength of sound in the frequency range $125 \mathrm{C} / \mathrm{s}$ to $250 \mathrm{c} / \mathrm{s}$. Moreover, as some recently published work ${ }^{1}$ has shown, that the acoustic impedance presented by the cabinet to the cone may change from a capacitance to an inductance at a frequency of which the wavelength of the sound is some seven times the maximum dimension of the enclosure. In some cases, therefore, the simple theory based on lumped constants may break down even below $100 \mathrm{c} / \mathrm{s}$. Reference to a table of absorption coefficients will show that the effectiveness of any sound absorbent lining of practicable thickness falls off below $500 \mathrm{c} / \mathrm{s}$. It will thus be seen that the frequency band within which standing wave effects are to be expected may extend over some two octaves in an important part of the audio-frequency spectrum.

A practical example will illustrate the foregoing discussion. An experimental closed cabinet (the so-called " infinite baffle" type) was required for a roin. cone unit; the resonance frequency of the unit with the cabinet volume was not to exceed $45 \mathrm{c} / \mathrm{s}$.

The suspension of the cone unit had been experimentally modified to give the maximum compliance consistent with mechanical stability, but even so a volume of $8 \mathrm{cu} . \mathrm{ft}$. was found to be required to meet the specification. For reasons connected with highfrequency distribution, the axis of the present loudspeaker was required to be about 3 ft . from the floor and this consideration led to the adoption of a design 4 ft . high.

Fig. I(a) shows the free-air axial frequency response obtained with this cabinet, the cone unit being driven by an amplifier of low output impedance. The curve was taken with the cabinet lined with

[^1]absorbent material of conventional type. Standing wave effects are nevertheless apparent at low frequencies, showing that the internal damping is inadequate.

## Damping by Partitions

The ineffectiveness of sound absorbent linings at low frequencies is due to the fact that all the absorbent material lies within a small fraction of a wavelength of the inner wall, which is, of course, a velocity node. Without motion of the air particles, or of the materia! itself, acoustic power cannot be absorbed and dissipated. Consideration of these facts suggests that the best position for damping material would be somewhere out in the space inside the cabinet, where the standing waves could be caught "on the wing." Theoretically, the material would have to be divided up into many parts, so disposed that at any one frequency there should be an absorbing element in the neighbourhood of a velocity antinode. In practice, however, such a complication has been found unnecessary and effective low frequency damping can be achieved by concentrating the absorbent material into one or two partitions strategically placed across the cabinet. Fig. I (b), for example, shows the response of the loudspeaker previously referred to when a single sheet of $\frac{1}{4}-\mathrm{in}$ carpet felt was stretched horizontally across the inside of the cabinet at right angles to the longest dimension; the reduction in standing wave effects can be clearly seen.

In this system ${ }^{2}$ of damping by partitions, the cab-


Fig. 1. Axial response below $500 \mathrm{c} / \mathrm{s}$ of ro-inch speaker in closed cabinet (a) inner surface lined with absorbent material in conventional manner; (b) with the addition of an absorbent membrane stretched horizontally across the cabinet at right angles to the longest dimension.

Fig. 2. Bridge circuit for motional impedance measurements.


2 Patent Application No. $2+528 / 49$
inet can be regarded as being divided into sections, each so small that any standing wave effects occur at high frequencies, at which the sound absorption of a lining of the conventional type is adequate. It may further be noted that those sections of the cabinet which are separated from the cone by one or more sound absorbing partitions receive little sound at the high frequencies and therefore require very little " acoustic treatment."
The principle of subdivision can he applied in a number of ways to suit different cases. The photograph, for example, shows part of an experimental vented cabinet constructed in two separate sections. The lower of the two sections houses the power amplifier and takes the form of a plinth on which the upper section stands. In the illustration the upper section is tilted to show the underside which has a window communicating with the plinth and covered with carpet felt.

## Impedance Test

Reference was made earlier to the strategic placing of absorbent partitions. This operation would appear at first sight to require facilities for measuring the overall frequency response of the loudspeaker. Fortunately, however, it is possible to make the necessary experimental adjustments by quite rough impedance measurements. Fig. 2 shows the circuit used for this purpose. $R_{1}$ is the resistance of the speech coil of the loudspeaker which is fed at constant current from an audio-frequency oscillator though a


Fig. 3. Typical response curves obtained with the circuit of Fig. 2. Curve (a) is for an undamped cabinet and (b) for the same cabinet with internal damping partition.


Fig. 5. Response of closed-cabinet loudspeaker for various valves of the effective damping resistance $R_{\text {s }}$.

Platively high resistance $\mathrm{R}_{2} . Z_{m}$ is the motional impedance of the loudspeaker, so that the voltage developed across $Z_{m}$ represents the back e.m.f. generated by the motion of the speech coil when driven by a constant current. Since we are dealing only with the frequency range in which the cone moves as a whole, this back e.m.f. is a measure of the cone velocity. The cone aperture is the only outlet from the cabinet; any internal resonances can only influence the sound output from the system by modifying the motion of the cone, and must therefore reveal themselves by irregularities in the curve of motional impedance against frequency.

In general, the resistance $\mathrm{R}_{1}$ will tend to swamp all but the largest of these irregularities so that the total impedance of the loudspeaker is not a very sensitive index of cabinet performance. In the circuit of Fig. 2, therefore, the effect of $\mathrm{R}_{1}$ is roughly balanced out. Resistance $R_{3}$ is made approximately equal to $R_{1}$, and the variable resistance $R_{4}$ approximately equal to $\mathrm{R}_{2}$, the whole circuit forming a bridge, the output of which is taken off through the isolating transformer T to a valve voltmeter. The primary impedance of $T$ should be high compared with the maximum speech coil impedance at low frequencies, a requirement which is generally easy to meet.

To make a test, the speech coil is temporarily prevented from moving, either by disconnecting the loudspeaker field supply or, in the case of a permanent magnet, by carefully inserting small wedges into the gap so as to grip the coil. Assuming complete clamping of the movement, $\mathrm{Z}_{m}$ will now be zero, and in this condition, the bridge is balanced at some convenient frequency, preferably in the $200 / 300-\mathrm{c} / \mathrm{s}$ region, by adjusting the value of $\mathbf{R}_{\mathbf{4}}$ and using the valve voltmeter as a null-indicator. The speech coil is then released and the bridge output, which now represents
the voltage across $Z_{m}$ (i.e. the back e.m.f. of the speech coil) is read direct on the valve voltmeter and plotted against the frequency of the oscillator. A typical curve of this kind is shown in Fig. 3(a) which was obtained in an ordinary live room with a completely undamped cabinet. The radiation efficiency of the loudspeaker is too low for any room resonances to affect the curve but the internal modes of the cabinet are clearly apparent. Fig. 3(b) shows how these modes are affected by the introduction of a felt partition.

It will be noted that in this test, the speech coil is being driven with constant force. If the cone were mass-controlled, i.e. if the compliance of the suspension were infinite and the effect of the cabinet nil the speech coil would move with velocity inversely proportional to frequency. In this hypothetical case, the curves of Fig. 3 would be straight lines having a slope of 6 db per octave; and it is sometimes convenient to use such a slope as a standard of reference. If desired, an equal and opposite slope can be introduced into the response of the test circuit (for example, by interposing in the input of the valve voltmeter a small series condenser). With this compensation, the valve voltmeter readings will be confined to a much smaller range, and may be plotted to a more open scale. Small irregularities in the curve may thus be better observed; and the reference line representing the condition of mass control becomes horizontal.

The method of test described above has been used successfully up to about $300 \mathrm{c} / \mathrm{s}$ and the accuracy, though limited by the difficulty of ensuring complete clamping of the cone movement, is sufficient to disclose the more troublesome internal cabinet resonances, so that these can be dealt with without recourse to free-air response measurements.

## Effect of Source Impedance

Returning to the response characteristic shown in Fig. I, it will be seen that although the fundamental resonance frequency of the system is at $45 \mathrm{c} / \mathrm{s}$, the acoustic output begins to fall away well above this frequency. For an efficient loudspeaker, driven by an amplifier of low output impedance, this type of response is in accordance with theory. The mechanism of the effect can be explained in the following way.

At the lower end of the audio-frequency band, the wavelength of the sound is large compared with the diameter of the cone, while the polar distribution of the sound output varies only slightly with frequency. It can be shown that in these circumstances the sound pressure produced is proportional to the product of the cone velocity and the frequency. Thus, constant sound pressure will be obtained if the cone moves with a velocity varying inversely as the frequency, while constant cone velocity will give a sound output falling towards the bass. If the cone unit is very efficient, its motional impedance at low frequencies will be high, or, in other words, practically the whole of the voltage across the loudspeaker terminals will be accounted for by the back e.m.f. generated by the motion of the speech coil. If, in addition, the impedance of the driving source is low, the back e.m.f. will be nearly equal to the source voltage, and must therefore remain nearly constant with frequency. In these circumstances, the cone must be moving with nearly constant velocity, and the sound output will therefore fall at the bass. Regarding the matter
from a slightly different viewpoint, we may say that the rise in impedance of the speech coil at lower frequencies produces an electrical mis-match which so reduces the efficiency of power transfer from amplifier to loudspeaker that, however efficient the electroacoustic power conversion of the system may be, constant overall efficiency cannot be obtained. The higher the magnetic flux density, the further up the frequency range does this effect extend, so that the
" infinite baffle" may give disappointing results with the better class of unit. This cloes not, of course, mean that high efficiency in a cone unit is a bad thing, but that in some cases it is difficult to make full use of it

It is sometimes suggested that one can never have too much electrical damping of a loudspeaker. The present case, however, appears to suggest the contrary, and in view of the general interest in the subject, it may be worth while to study the electro-acoustic system from yet another view-point.

Fig. $4^{(a)}$ is a simplified electrical equivalent circuit of the acoustic system of the loudspeaker. For reasons which will appear later, the circuit elements are not placed in quite the usual order. $L_{0}$ represents the cone mass plus the effect of radiation reactance, $\mathrm{R}_{\mathrm{U}}$ the small radiation resistance (which unlike most other resistances, varies with frequency), $C_{C}$ and $C_{U}$ respectively the acoustic capacitance of the cabinet volume, and the equivalent capacitance of the cone suspension. $R_{\mathrm{s}}$ is a very important quantity which, nevertheless, is frequently omitted from these equivalent circuits. It represents the effect of the electrical circuit of the loudspeaker and driving amplifier reflected into the acoustic circuit of the loudspeaker in a similar way to that in which the motional impedance is transferred into the electrical circuit. At low frequencies,

$$
\mathrm{R}_{\Delta} \propto \frac{\mathrm{B}^{2}}{r_{a}+r_{c}}
$$

where B is the flux density in the gap, $r_{a}$ is the output impedance of the driving amplifier (assumed resistive) and $r_{c}$ is the d.c. resistance of the speech coil. Any internal frictional resistance in the cone suspension
system can, for simplicity, be lumped with $\mathrm{R}_{8}$ which, together with an equivalent constant voltage generato $\mathrm{E}_{\mathrm{s}}$, represents the source of power.

It has already been remarked that at low audio frequencies, the pressure response of the loudspeaker is proportional to the product of cone velocity and frequency. In the equivalent acoustic circuit of Fig. 4 the current $I_{U}$ represents the alternating air current produced by the cone and is thus proportional to the cone velocity. Hence the acoustic response of the loudspeaker will be proportional to $\mathrm{I}_{\mathrm{v}} f$. The voltage $\mathrm{E}_{\mathrm{L} U}$ is $2 \pi f \mathrm{~L}_{\mathrm{U}} \mathrm{I}_{\mathbf{U}}$. Thus $\mathrm{E}_{\mathrm{L} U}$ is proportional to $I_{\mathrm{u}} f$ and the variation in this voltage with frequency for any one value of $L_{\mathrm{U}}$ gives the frequency response of the loudspeaker. Over the frequency range for which the equivalent circuit is valid and the wavelength is large compared with the size of the cone, the loudspeaker can be reduced, as far as frequency response is concerned, to a half-section, high-pass filter working into open circuit. Small values of $\mathrm{R}_{\mathrm{s}}$, result ing from low flux density or high amplifier output impedance give a resonance peak and bad transient response, while large values of $\mathrm{K}_{\mathrm{s}}$, corresponding to high flux density and low amplifier output impedance can give a serious loss of bass.

The range of possibilities is illustrated by the three curves shown in Fig. 5, showing the variation of $\mathrm{E}_{\mathbf{L U}}$ for a resonance frequency of $45 \mathrm{c} / \mathrm{s}$ with different values of $\mathrm{R}_{\mathrm{y}}$. Curve 3 will be recognised as approximating to the curve of Fig. $\mathrm{I}(\mathrm{b})$, and it will be seen that when standing wave effects are disposed of, the response of the loudspeaker approaches the form predicted by simple theory. Curve 3 corresponds to a $Q$ of 0.46 , only slightly less than the figure of 0.5 required for critical damping. Whether critical damping is really necessary is another matter, to which we shall have occasion to refer later; but in the meantime it should be noted that this condition can only be achieved at the expense of a drooping frequency characteristic.
(To be concluded)

## Standardized Components

THE Admiralty announce that their range of hermetically sealed transformers and chokes are the first type to qualify as fully Inter-Service Type-Approved Standards. The design comprises "C" core assemblies

accommodated in deep-drawn steel cases, and owing to the many technical improvements which have been incorporated, the new transformers and chokes are as much as $40 \%$ smaller and give up to $50 \%$ reduction in external magnetic field compared with their counterparts using orthodox laminations.

The contour for each size has been carefully determined to provide the maximum strength, and construction is such as to withstand the severest conditions of vibration and shock, as well as extremes of climate.

Thirty-two sizes are available covering range of power transformers from $5-\mathrm{VA}$ to $2-\mathrm{kVA}$, also a range of audio-frequency transformers and power chokes to suit all normal requirements. Thirty-one of these sizes have tapped fixing holes at each end, thus permitting upright or inverted mounting at will.

Although the design of this range is Admiralty property, it may be usecl freely for commercial purposes by any manufacturer who so desires. Where firms do not wish to tool the range for themselves, they can obtain supplies of cases, internal fittings, etc., from commercial sources. More than twelve firms have already taken advantage of this arrangement, and as their production increases so a greater flow of these standardized transformers and chokes will reach the home and export markets.

# Pickup Input Circuits 

Compensating for 78 and $331 / 3$ r.p.m. Recording Characteristics

By R. L. WEST, b.Sc., A.m.Brit.I.r.e., and S. KELLY

MUCH disappointment can be avoided by a simple understanding of the principles underlying the design and selection of input arrangements for standard 78 r.p.m. and $33 \frac{1}{3}$ r.p.m. long-playing records. An exhaustive treatment is not intended, but it is hoped that this article will help the beginner to avoid the commoner pitfalls.

Most pickups fall into two main types-crystal (or piezoelectric) and magnetic-the latter covering ribbon and moving coil, as well as moving-iron armature and " variable reluctance" types. Crystal pickups are always of high-impedance; they are thus suitable for more or less direct connection to a grid cjrcuit. Magnetic pickups are sometimes wound with a large number of turns of wire to generate the relatively large voltage required for the grid circuit; this can introduce electrical resonance (of self capacity and inductance of coil) unless great care is taken in the design. For high-fidelity pickups it is normally more convenient to use fewer turns (only one in the ribbon) these produce very small e.m.f.'s but are capable of delivering a much larger current. Since it is voltage and not current that matters at the grid circuit a suitable step-up transformer is normally used. The following remarks assume the use of a transformer where necessary and apply mainly to the high-impedance (secondary) side.

Effect of Load.- The input impedance of a valve is usually very high compared with the generator impedance and can be neglected; the value of the grid leak will therefore be dictated by the load impedance requirements of the pickup.

Magnetic pickups have internal impedance which is principally L and R in series, whereas crystal types

Fig. 1.-Effect of resistance on output from magnetic and piezoelectric pickups.

(a)

(D)
are principally $C$ and $R$ in series and $R$ is usually very small. Fig. I shows the effect of load resistance and assumes constant output voltage on open circuit. In each case the dotted characteristic represents the effect of a lower value of load resistance.

Incidental Capacitance.-(i) Screened lead between pickup (or transformer secondary) and amplifier.Again the internal impedance has to be introduced (see Fig. 2). By way of simplification the small resistive component has been omitted from the crystal case but should be included in the event of using very high cable capacities. The slight peak shown in the magnetic case is seldom noticeable since the internal R and the external R (not shown) damp it very thoroughly.
(ii) Screened lead after the volume control.-This is a common trap for beginners and is often overlooked by those who should know better! The Fig. 2 (c) shows the effective circuit, and the resulting top loss with intermediate settings of the volume control, which can be very considerable, particularly if a high-value volume control is used. The effect disappears as the slider approaches the "bottom end" and turns into an example of the previous type when the slider approaches the "top end." The use of a compensating condenser $\mathrm{C}_{6}$ as shown sometimes helps a little, but note that a capacity varying between $C_{L}$ and $C_{L} / 2$ is now permanently across the input, and only when the slider is half way (electrically) is the "compensation" correct! Far better, if enough gain is available, to use circuit of Fig. 2 (d). Here the very low output resistance will "swamp" the capacity of most normal screened lead requirements.

Pickup Resonances.-Low resonances, say under $1,000 \mathrm{c} / \mathrm{s}$, include those due to the tone arm torsional resonance and the effective mass of the whole pickup resonating with the armature mounting compliance. If these are excessive within the working range, the pickup can be considered unsatisfactory, since it will be found that very heavy tracking pressures are necessary to keep the needle in the groove at these frequencies, with consequent increase in record wear. Electrical correction is no remedy.
The most noticeable high-frequency resonance is where the stylus and/or armature flexes. This ranges from about $3,500 \mathrm{c} / \mathrm{s}$ in the older pickups to well above audibility in some modern ones. The "height" of the resonance varies, from 12 db or more in the case of an undamped system, down to a barely perceptible rise if sufficient damping is added and in the right place. This is the resonance which, if within the audible range, considerably augments needle scratch. Above this resonance, the output usually falls off very rapidly (see Fig. 3a).
A simple treatment for magnetic types is to use a rather lower load than normal, and so produce top


Fig. 2.-Illustrating the effect of external capacitance on the response of magnetic and crystal pickups.
attenuation at the rate of 6 db per octave. This improves the general balance-Fig. 3(b)-and assists the electromagnetic damping.
A more elaborate circuit uses a tuned series filter -Fig. 3 (c). K is seldom needed as the pickup impedance is usually sufficient. As a rule $\mathrm{K}_{2}$ can likewise be dispensed with since there is some resistance in the inductance. It is better to over-emphasise the correction in order to reduce to a minimum surface noise due to the armature resonance. "L" can be an air-cored or dust-cored choke of between $\frac{1}{4}$ and I henry. An ordinary laminated core usually exhibits a marked change of inductance with signal strength at these very low operating levels.

The older crystal types usually had an overall output of the type shown in Fig. 3 (d). Here it will be seen that the overall balance is sufficiently good for average domestic use.

Results can be improved by a tuned filter, a parallel-tuned (rejector) circuit is the simplest to use. In Fig. 3 (e), "L" would be the same component as in the magnetic case.

Choice of Load.-In general, the higher the load resistance, the greater the voltage developed by the pickup. For magnetic types the load can be several megohms if the top resonance frequency is very high, the grid circuit capacitance low and no top attenuation desired at this point. The makers' recommendations will hāve taken these factors into consideration.

For the older crystal types it is usually necessary to use a load under one megohm in order to attenuate the bass response somewhat-unless one is trying to get bass from a small cabinet! Values of $\frac{1}{2}$ megohm to 100,000 ohms are most common.

Hum-Causes and Cure.-There are two main sources of hum-by induction from an alternating magnetic field such as from the mains transformer or gramophone motor, and by electrostatic induction from wiring and components usually connected to the mains or other high voltage a.c. sources.-

Magnetic hum introduced into the leads and wiring represents only a very small e.m.f., since only one complete turn is involved. This can be troublesome when a step-up transformer follows the lead in question, how troublesome depends on the trans-
former turns ratio and the e.m.f. generated by the pickup.

The best treatment is to twist tightly these lowimpedance leads all the way from the pickup coil itself right up to the transformer input terminals. For the secondary connections, ordinary screened leads are sufficient, but it is advisable to keep all these leads as far from stray magnetic fields (including heater wiring) as possible. The transformer hum problem is dealt with later. On the high-impedance side the magnetic and crystal types experience mainly hum from electrostatic induction. The cure is simple-just plain good screening everywhere, and this precludes mains switches on volume controls, unless they are well shielded.

Rumble.-This consists usually of vibrations originating from the motor, with the main com-

Fig. 3.-" Top " resonances in magnetic and crystal pickups.

ponents between about 5 to $30 \mathrm{c} / \mathrm{s}$. Magnetic pickups are seldom troubled with rumble since their output is proportional to relocity which falls with frequency for a given amplitude-hence very little output occurs at these low frequencies.

Crystal pickups, on the other hand, usually show up the motor deficiencies on this score, since the output voltage is proportional to amplitude. In a recent design (Acos GP2o) a velocity type characteristic has been introduced below about $30 \mathrm{c} / \mathrm{s}$ and the trouble is considerably reduced.

Where necessary, a simple high-pass filter, such as that shown in Fig. 4, gives useful rumble attenuation without spoiling the bass response. This attenuates at $12 \mathrm{db} /$ octave and gives a more rapid rate of fall at small values of attenuation than the more usual circuit, having equal capacities and equal resistances. Using the circuit shown, values of


Fig. 5.-Top boost controls (a) with variable cut-off frequency, (b) with variable attenuation slope.

Fig. 6.-Principal commercial recording characteristics plotted on the basis of equal velocity at $1,000 \mathrm{c} / \mathrm{s}$. A, standard 78 r.p.m. ; B, Decca FFRR, 78 r.p.m. ; C, Decca long playing.

$\mathrm{C}=0.01 \mu \mathrm{~F}$ and $\mathrm{R}=0.5 \mathrm{M} \Omega$ have proved satisfactory.
Pickup Transformers.-(a) Step-up ratio.-In general, as high a step-up ratio as practical is used. Suppose the grid leak-cum-load has been chosen as 1 MS, then the actual load on the pickup will be $10^{6} / \mathrm{N}^{2}$ where N is the turns ratio. The larger N , the smaller is the load, so one must not make N too large by trying to increase the output voltage too far, or severe top loss will occur. If the d.c. resistance of the pickup is, say, 25 ohms, then its impedance at $1,000 \mathrm{c} / \mathrm{s}$ we know must be more than this, say 50 ohms, and the load should be much greater still, say, 100 ohms if minimum top loss is desired. This makes $\mathbf{N}$ equal to 100 . If top loss is definitely desired $\mathbf{N}$ could be increased to, say, 200. This will not quite double the output voltage since the lower load will cause a larger voltage drop (at all frequencies) in the resistive component of the pickup impedance.
(b) Choice of primary turns.-For any given ratio and core, two few primary turns, i.e., low inductance, will lead to loss of bass.

At the lowest frequency involved the primary reactance must be equal to or greater than the pickup impedance or load, whichever is the larger. A rough and ready rule, which is liable to err only on the generous side is to put the same number of turns on the primary as there are in the pickup coil, be it one or many.
(c) Core Material.-In the interests of minimum distortion (due to hysteresis), at the very low signal levels involved, a nickel alloy such as Mumetal is by far the best material. Not much will be needed, a $\frac{1}{2}$-in square core section should be ample for any design.
(d) Screening.-Enclosure in any earthed metal can will look after the electrostatic component. From the magnetic point of view the best method is to select a spot as far as possible from the mains transformer, smoothing choke and motor. Use a Mumetal can, which should be earthed, care being taken to prevent the transformer core from touching the can, and remembering that the magnetic properties of Mumetal deteriorate if the material is stressed by cold working in any way. Any residual (magnetic) hum can be reduced by orienting the transformer in the can, or the can as a whole. In severe cases, a second Mumetal can, to enclose without touching the first, may be necessary.

Mechanical Feedback.-On occasions, when the loudspeaker is in the same cabinet as the turntable, mechanical feedback will occur when the pickup stylus is in contact with the record. This is usually due to fimsy cabinet construction or to attempting a very large low-frequency output. Each case must be treated on its merits, but trouble of this nature emphasises the desirability of a separate speaker. When this is not practicable cases may be dealt with by rubber or felt mounting for the whole baseboard, stiffening the baseboard, tightening or slackening slightly the motor mounting, or even reducing the bass response at the extreme low frequencies.
Simple Pickup Measurements.-Very little apparatus is necessary to carry out useful checks on frequency characteristics. A standard frequency record, preferably the type with bands of fixed frequency ranging from, say, $30 \mathrm{c} / \mathrm{s}$ up to $14,000 \mathrm{c} / \mathrm{s}$ or more, a fixed resistor of 5 to ro-watt rating equal to the nominal speaker load, an a.c. voltmeter (rectifier type) of range o to 5 V or 0 to 10 V .

Most modern amplifiers employ sufficient feedback
to be virtually flat over the audio range, so, with a resistive load in place of the speaker and the voltmeter across that, they make a very nice valve voltmeter, provided the volume control is not disturbed after the initial setting.

A response curve can then be obtained quite easily by converting voltage ratios to the voltage at, say, r, $000 \mathrm{c} / \mathrm{s}$ into decibels by the usual formula: Decibels $=20 \times$ log. ratio-or by referring to decibel tables or abacs if these are available.

Be careful not to overload the amplifier when taking these readings. Knowing the maximum power output of the amplifier and remembering that Watts= V. ${ }^{2} \mathrm{R}$ calculate the highest reading V you can allow to be seen on the voltmeter.

In the absence of an LP test record, the circuit can be checked satisfactorily using a 78 -r.p.m. standard frequency record, run at 78 r.p.m. and using the correct stylus. To the readings obtained, when converted to decibels, add the bass-cut figures quoted on the record, then the final curve should look like the inverse of curve C in Fig. 6 if equalization is correct. This method is quite accurate except for the top resonance, if any.

Controls for a Pickup.-Two controls are really suf-ficient-a top attenuator, preferably switched in 4 or 5 stages, to cover age, origin, and condition of 78 r.p.m. records, and a changeover switch to effect the major 78 -LP change. A three-position switch is useful, in the form 78 -LP-Radio. The more ambitious might like to expand it to :-78 NORMAL- 78 FFRR -LP-Radio, but the extra top of the Decca FFRR can be dealt with quite adequately by the normal top control.

Two top cut circuits are shown. Fig. 5(a) is the conventional one with $6 \mathrm{db} /$ octave attenuation, starting higher or lower in the scale according to the capacity chosen. With the values given attenuation starts, according to the switch position, at frequencies in the neighbourhood of 10,6 and $3 \mathrm{kc} / \mathrm{s}$ and for really bad records at about $300 \mathrm{c} / \mathrm{s}$. Fig. 5(b) varies the slope from 5 to $20 \mathrm{db} / 0 \mathrm{ctave}$, with a little variation of the starting point, which is in the region of $1,000-$ 2,000 c/s.

A $78 / \mathrm{LP}$ changeover is suggested, rather than using the top and bass controls; this enables the changeover to be made with a single operation. Further, exact equalization of LP recordings is not possible with simple cut / boost controls, and it is in any case desirable that the whole of the variable top and bass control range should be available for special conditions.

78 r.p.m.-(i) The recording characteristic.-Fig. 6 shows (A and B) the two recording characteristics produced in this country in terms of velocity against frequency.
(ii) Correction circuits.-The magnetic types require a bass-lifting circuit of the type shown in Fig. 7(a). In reality it " attenuates-everything-but-the-bass," a matter of io times for both sets of values given, so that adequate gain must be available in the amplifier.

The circuit of Fig. 7(b) for the crystal type is similar in this respect. With the older crystal types it must be used with discretion, though, on account of the rather large high-note resonance. An elaboration of this circuit which was recommended for use with the Acos $G_{12}$ is shown in Fig. 8. This pickup followed closely the theoretical amplitude operation of piezo crystals.

With the later types of crystal pickup, such as the Acos GP2o, the high-frequency response does not
follow this law, but has an internally-compensated response which approximates to a velocity law at high frequencies when terminated by a resistive load. For those who would like to improve the response, a circuit is shown in Fig. 9.

| $R_{1}$ | $R_{2}$ | $R_{3}$ | $C$ |
| :---: | :---: | :---: | :---: |
| $100 \mathrm{k} \Omega$ | $10 \mathrm{k} \Omega$ | $\geq 500 \mathrm{k} \Omega$ | $0.05 \mu \mathrm{~F}$ |
| $200 \mathrm{k} \Omega$ | $20 \mathrm{k} \Omega$ | $\geq 1 \mathrm{M} \Omega$ | $0.05 \mu \mathrm{~F}$ |


(d)

(b)

Fig. 7.-Simple compensating circuits for 78-r.p.m. recordings (a) magnetic (velocity) pickups, (b) crystal (amplitude) pickups. The dotted curves indicate voltage output before correction.


Fig. 8.-Correction circuit for the Acos GPiz pickup on 78 r.p.m. records.

Fig. 9.-Correction circuit for Acos GP2o pickup on 78 r.p.m. recordings.


Fig. 10.-Complete compensating circuit ( 78 r.p.m.) for magnetic (velocity) pickups.


None of these circuits include top correction for the difference between standard and FFRR characteristics, but this will be covered by the suggested top control

A complete circuit for a magnetic pickup is shown in Fig. ro, and for an Acos GPzo in Fig. II. This latter includes the anti-rumble circuit of Fig. 4.

Long-playing ( $33 \frac{1}{3}$ r.p.m.).-The successful adaptation of standard pickups for microgroove recording i dependent on the recognition of several factors. If the pickup will not track standard 78 -r.p.m. test recordings satisfactorily at 14 grams or less, it is improbable that the same pickup (with a correct radius


Fig. II.- Complete compensating circuit (78 r.p.m.) or Acos GPzo pickup.

Fig. 12.-Output of high-impedance Decca Model D2 (3-pin type) moving-iron pickup on $33 \frac{1}{3}$ r.p.m. test record; $A$, without corre tion ; $B$, with equalizer circuit shown inset.


Fig. 13.-Low-impedance Leak moving-coil pickup and transformer on $33^{\frac{1}{3}}$ r.p.m. test record ; A, without correction ; B, with equalizer circuit.
stylus, of course) will track long-playing records at 7 grams. The tracking problem is not only important at the low frequencies, but also at the extreme high frequencies, where the velocity of the microgroove recording approaches that of the standard record, although the tracking weight of the pickup is considerably less and mechanical impedance of the armature is rising rapidly.

Assuming, however, that the pickup is satisfactory in this respect, there is no reason why it should not give satisfactory results on microgrooves, providing it is correctly equalized. It should be noted that the effective resonance frequency of the armature system is usually decreased by about half an octave on microgroove compared with standard records, so that if any resonance is at all apparent in the upper register on standard records it will, in generabl, be more prominent and at a lower frequency on microgrooves.

Frequency correction.-As examples, two highfidelity magnetic pickups and a similar type crystal pickup are presented herewith. The open-circuit response characteristic of the Decca Model $\mathrm{D}_{2}, 3$-pin type, magnetic pickup is given in Fig. i2(A). It will be seen that this response, in the mid and lower registers, approximates to the recording characteristic, Fig. $6(\mathrm{C})$, but in the higher frequencies rises rather more steeply because of the lowered resonant frequency of the armature system. The electrical network, shown inset, corrects the response of the pickup and gives the overall response shown in curve $B$. Although this final response is not in the " straight line from d.c. to infinity " beloved by the pedants, it is well within $\pm 2 \mathrm{db}$. The components in question were radio-tolerance units. It may be pointed out that the 500 pF terminating condenser and the 4,ooo-!? resistance may have to be varied with individual pickups to get a satisfactory balance between the middle and upper frequencies. This equalizer has been successfully used with a variety of pickups of up to 5,000 ohms impedance (connected direct or taken on the secondary of the coupling transformer) which normally require a load resistance of quarter to half megohm.

The best of the moving-coil pickups show a resonance of at least $20 \mathrm{kc} / \mathrm{s}$ on standard $78-\mathrm{r}$.p.m. records and even when played on microgroove records the resonance is seldom lower than 15 or $16 \mathrm{kc} / \mathrm{s}$ and the pickup response is very nearly that of the record. With care the low-frequency resonance can be below $30 \mathrm{c} / \mathrm{s}$ and the low-frequency response will also be

Fig. 14. Cosmocord GPzo crystal pickup on $33{ }^{1}$ r.p.m. test record ; A, without correction ; B, with " bridged $T$ " equalizer circuit.

very nearly that of the recording characteristic. The Leak moving-coil pickup and its transformer are shown as being representative of this type of instrument. The open-circuit response is given in Fig. $13(\mathrm{~A})$; when connected with the appropriate equalize network the response shown at B is obtained. The high-frequency "roll-off" is controlled by the 0.015 $\mu \mathrm{F}$ condenser ; decreasing it to o.or $\mu \mathrm{F}$ will increase the ro-kc/s response by about 4 db . This condenser can be adjusted to meet individual requirements. If the low-frequency end is considered excessive a condenser can be inserted between the transformer secondary and the $44,000 \Omega$ resistor. A value of $0.25 \mu \mathrm{~F}$ will give a reduction of about 6 db at $50 \mathrm{c} / \mathrm{s}$. The "roll-off" at low frequency can be adjusted to suit conditions by varying the value of this condenser, lower values increasing the attenuation

The case of the crystal pickup is shown in Fig. 14,
the unequalized response being shown at A and the equalized at $B$. It will be seen that a modified "bridged T" network is used, and, within reason, the equalizing is independent of the pickup impedance. In all cases the terminating resistance should be 0.5 MQ . If the input impedance of the amplifier is other than this value, a simple potential-divider matching arrangement should be used.
It may be found, especially with cheaper type turntables or units that have been modified from 78 r.p.m., that motor rumble is excessive. Should this be the case, the high-pass filter unit described earlier may be used successfully, but should be connected between the equalizing unit and its load resistance.

In conclusion, the authors are indebted to Messrs. Decca Radio and Television, Cosmocord and H. J. Leak \& Company for information regarding characteristics of records and pickups.

# SHORT-WAVE CONDITIONS 

## October in Retrospect : Forecast for December

By T. W. BENNINGTON (Engineering Division, B.B.C.)

DURING October the average maximum usable frequencies for these latitudes increased very considerably during the daytime, and decreased considerably during the night. These variations were in accordance with the normal seasonal trend.
Daytime working frequencies for long-distance communication were fairly high, though not so high as had been expected. The $28-\mathrm{Mc} / \mathrm{s}$ band was sometimes, but not often, usable to the U.S.A., though it was frequently usable in more southerly directions. The failure of the working frequencies to increase as much as was expected was probably due, in part, to the large amount of ionospheric disturbance, and in part to the rapidly decreasing solar activity. Night-time working frequencies were also low, being generally below $9 \mathrm{Mc} / \mathrm{s}$. There was a small decrease in the rate of incidence of Sporadic E , and not much communication on high frequencies occurred by way of this medium.
Sunspot activity was, on the average, slightly higher than during September, but the general level has now fallen by well over one-third since sunspot maximum. Several severe ionospheric storms occurred during the month, the most disturbed periods being ist to 8 th, $14^{\text {th }}$ to 18 th, 23 rd to 24 th and 28 th to 3 1st. No Dellinger fadeouts have, as yet, been reported.

Forecast.-There may be a small decrease in the daytime m.u.fs for these latitudes during December, as compared with those for November. Night-time m.u.fs should also decrease and perhaps reach their lowest values for the present winter season.

As a result of these variations, the long-distance work= ing frequencies should be rather high by day, though the higher frequencies will be usable only for relatively short periods daily. It is doubtful whether $28 \mathrm{Mc} / \mathrm{s}$ will be regularly usable over most circuits, though it may well be so on those running in southerly directions. The medium-high frequencies will provide the main means of daytime communication in most directions, and at night 6 to $7 \mathrm{Mc} / \mathrm{s}$ should be the highest regularly usable frequencies.
Sporadic $E$ is not likely to be very prevalent and medium-distance communication on high frequencies by way of this medium should be very infrequent. Iono-
spheric storms are not, as a rule, particularly common during December, but those which do occur are likely to be troublesome, particularly at night.
The curves indicate the highest frequencies likely to be usable over four long-distance circuits during the month.


Aerials for Centimetre Wavelengths (Modern Radio Technique Series). By D. W. Fry and F. K. Goward. Pp. $172+x$, figs. 65. Cambridge University Press, 200, Euston Road, London, N.W.i. Price 18s.

THIS is a book of interest primarily to engineers concerned with the design and development of specialized aerials for radar applications at centimetric wavelengths. The authors themselves were concerned at the Telecommunications Research Establishment in much of the early design of aerials of this type. The treatment is not elementary, and a considerable knowledge of transmission lines, wave-guides, and general aerial and radiation theory on the part of the reader is assumed. Nevertheless, the mathematical aspect is not unduly laboured, and inathematical expressions are confined to the minimum necessary for the purpose of theoretical design.

It is in no sense a practical designer's handbook. It describes and catalogues the various types of aerials which have been used or are used for particular radar work, and it discusses in considerable detail the analytical methods of designing a radiating system so as to obtain a specified radiation pattern. Generally a centimetric aerial comprises two principal parts-a primary radiator, such as a dipole or an open-ended wave-guide, which feeds the energy to the secondary radiator, the latter being a reflector, a lens or a horn: primary and secondary radiators are dealt with in separate chapters, which are sub-divided according to the various distinct types.

A large part of the work is concerned with means of scanning, that is, causing the radiated beam to explore continuously a particular area in space. It is shown to be a valuable feature of centimetric aerials that scanning can usually be accomplished at the primarv radiator which leads to simpler mechanics and higher scanning speeds than if displacement of the secondary radiator were necessary.

The close resemblance between centimetric work and optical theory is well brought out, and the application of optical methods to the design of centimetric reflectors and lenses is described. Both dielectric and metal plate type lenses are treated, and a note on errors and aberrations in optical systems is given in an appendix.

There are liberal references in the text to relevant literature, all the references being grouped in the form of a valuable bibliography at the end of the book. Printing and production are alike excellent, and the book is a worthy addition to the publishers' " Modern Radio Technique" series.
C. G.

Super-Regenerative Receivers. By J. R. Whitehead, Ph.D., A.M.I.E.E. Pp. I69 + xiii. Cambridge University Press, 200, Euston Road, London, N.W.I. Price 2IS.

## B

 EFORE the war the super-regenerative receiver had gained a reputation for being a tricky and unpredictable device that sometimes gave wonderful results. Published information was considerable in quantity, but generally evasive when it came to clear-cut questions such as : how its signal/ noise ratio compared with that of a superheterodyne; how one would design it for a given selectivity; how much gain could be obtained, and what factors determined it; what was the optimum quench frequency and waveform; how the shape of the frequency response curve compared with those of other types of receiver; and so on.The answers to these and other questions are given in clear and definite terms in the book under review-the first to be devoted exclusively to the super-regenerative receiver. Although it has resulted from work that was done in developing radio and radar equipment for the war, it is by no means a one-sided account of specialized
development. On the contrary, the author has started by getting down to the essential core of any superregenerative circuit; in fact, of any oscillator-a parallel combination of inductance, capacitance, and conductance. The special feature of super-regenerationperiodical variation of the conductance each side of zero -is then considered. The treatment is therefore perfectly general, and the influence of particular conditions and modes of operation are only brought in when the more general results have been fully established.
The treatment is mathematical and makes use of fairly advanced techniques, but, even if the reader is insufficiently adept to follow all the steps, he should have no difficulty in making use of the results, for the author has arranged his work admirably and has expressed the results clearly in words and in formulae that can be directly applied in design. An appendix contains valuable graphical design data, and the final chapter is devoted to a representative selection of practical circuits, complete with component values.

Most of the pre-war knowledge and use of the superregenerative receiver were in its logarithmic mode, in which oscillations are allowed to build up to saturation, resulting in strong a.g.c. action, and distortion of deep modulation. So it is only reasonable that special prominence should be given to the linear mode, in which at the cost of some complication of the equipment the distostion is avoided and greater selectivity obtained. The differences between this and the logarithmic and selfquenching modes are adequately coverd.

The proof of this attractive-looking pudding was in the remarkable reliability and consistency in pertormance of 200,000 IFF receivers (to mention only one item of super-regenerative equipment) during the war. The approximations which the author has had to make in order to get anywhere at all with his mathematical treatment seem to be fully justified by the results, within the limits laid down. Confidence in the diagrams is reinforced by a number of excellent oscillograms from actual receivers, in which the various conditions are clearly exemplified.

Errors appear to be confined mainly to incidental results of last-minute changes in the book; for example, what appears as Appendix 2 was presumably at one time No. 3, for several of the references are so numbered. Similarly with Fig. 3.3. And a paper which on p. 67 is described as " in the Press" has, by the time one reaches the Bibliography, been published. These are trifling slips. One would wish that every technical book treated its subject so satisfactorily.
M. G. S.

How to Become a Radio Amateur is a 58-page American publication, now in its eleventh edition, and its aim is to show the beginner how to make a start in amateur transmission. It opens by giving a very brief account of basic radio theory

The main body of the book is devoted to the choice of apparatus for the beginner with emphasis placed on construction as being the surest way of getting to know how things work. The most ambitious receiver described is a 4 -valve communications superhet. V.H.F. apparatus is included.

The book, which is published by the American Radio Relay League, West Hartford, 7, Connecticut, U.S.A., is well illustrated with diagrams and photographs and in the U.K. is obtainable through the Radio Society of Great Britain, New Ruskin House, Little Russell Street, London, W.C.I. It is delivered direct from the U.S.A. in from four to five weeks and the price is 4 s .

# Variable Filter Tuning 

(Conchuded from p. 358 of the previous issue)

2.-Design Procedure for R.F. and I.F. Stages

By A. B. SHONE, B.Eng., A.M.I.E.E.

IN the earlier part of this article we established in Fig. 3 the general shape of the receiver response characteristic at which we intend to aim.
It is possible to obtain such a response in the longwave band by any of the three alternative circuits which were given in Fig. 9. Considering the case of the long-wave Droitwich station, the response required is shown in Fig. ro, i.e., sensibly flat from $195 \mathrm{kc} / \mathrm{s}$ to $205 \mathrm{kc} / \mathrm{s}$ and heavily attenuated at igr $\mathrm{kc} / \mathrm{s}$ and below and at $209 \mathrm{kc} / \mathrm{s}$ and above. Using filters, this would require sections introducing a heavy attenuation at 1.02 times the cut-off frequency (i.e., $209 / 205 \mathrm{kc} / \mathrm{s}$ ). As will be seen in the curves given later, this is quite practical.
In the medium-wave band it is not really practical to obtain such characteristics directly by any of the three methods compared, and a superheterodyne type of circuit becomes almost essential.
It is in choosing a suitable intermediate frequency that the filter-tuned receiver begins to depart from normal practice. If the delineation of the response is to be divided between the high-frequency filter and the intermediate-frequency filter, then it is reasonable, if not axiomatic, to suppose that the minimum number of sections will be required when the work is shared equally between the two filters.

Referring to Fig. II: If $f_{k}$ represents the carrier frequency of the wanted transmission and $f_{i}$ and $f_{l}$ the extremes of its sidebands, if $f_{p}$ represents the carrier of the second channel and $f_{o}$ and $f_{s}$ its objectionable sidebands and if $f_{c}$ represents the intermediate frequency with $f_{b}$ and $f_{d}$ being the sidebands it is desired to receive unattenuated and $f_{n}$ and $f_{0}$ being the adjacent stations carriers it is required to reject, then we can set down this equation if the two filters are to be of equal sharpness :-

$$
\frac{f_{o}}{f_{l}}=\frac{f_{g}}{f_{d}}
$$

or

$$
\begin{gathered}
\frac{f_{k c}+2 f_{c}-5 \mathrm{kc} / \mathrm{s}}{f_{k_{c}}+10 \mathrm{kc} / \mathrm{s}} \\
=\frac{f_{c}+9 \mathrm{kc} / \mathrm{s}}{f_{c}+5 \mathrm{kc} / \mathrm{s}}
\end{gathered}
$$

Taking the worst positio in the medium-wave band i.e. when $f_{k}=1,500 \mathrm{kc} / \mathrm{s}$, w get $f_{c}=56 \mathrm{kc} / \mathrm{s}$ approxi mately. This intermediate frequency will require filter: whose sharpest sections introduce a heavy attenuation
 at 1.06 times their cut-oft
intermediate frequency slightly. If it is increased to $80 \mathrm{kc} / \mathrm{s}$, the sharpest sections in the high-frequency filter become r.oQ (i.e., $\frac{1,500 \mathrm{kc} / \mathrm{s}+160 \mathrm{kc} / \mathrm{s}-5 \mathrm{kc} / \mathrm{s}}{1,500 \mathrm{kc} / \mathrm{s}+10 \mathrm{kc} / \mathrm{s}}$ ) while the sharpest sections in the intermediate frequency filter are correspondingly reduced to 1.047 (i.e., $\frac{80 \mathrm{kc} / \mathrm{s}+9 \mathrm{kc} / \mathrm{s}}{80 \mathrm{kc} / \mathrm{s}+5 \mathrm{kc} / \mathrm{s}}$ ).

However, as will be seen below, this process should not be carried too far as the very sharp filter sections become relatively inefficient.

Considering now the actual filter sections available, the filters can be built in the form of high- and lowpass filters in tandem, the former having a lower cut-off frequency than the latter. Taken together, they jointly form a bandpass filter. This arrangement, although requiring a few more components than conventional bandpass filters, is preferable as the component values are in general much more convenient.

The sections we can use are tabulated in Fig. i2, together with the relevant formulac for calculating component values expressed in terms of R the terminal impedance of the filter, $f_{c}$ the cut-off frequency, and a term $m$. The latter term is evaluated by the formula $m=\sqrt{\frac{a^{2}-1}{a}}$ where $a$ is the ratio of the frequency of maximum attenuation to the cut-off frequency of the filter section. We can use either the " $\pi$ " form of the section of the " $T$ " form, whichever is
the more convenient. As we want to keep the variable condensers down to a minimum, we shall in general use the " $T$ " form of the low-pass filter and the " $\pi$ " form of the high-pass filter.

In each case there are two main sections, the prototype sections and the derived sections. (In each case the filter ends in half a derived section.) The prototype section begins to attenuate at the cutoff frequency and contimues with steadily increasing attenuation all the way to infinite frequency in the case of the low-pass filter (or to zero frequency in the case of the high-pass filter). The derived sections, however, are arranged to give their maximum attenuation at some frequency between infinite frequency (in the case of the low-pass filter or zero frequency in the case of the high-pass filter) and cutoff frequency. Between these limits we can make the frequency of maximum attenuation of a derived section what we like, and we describe the section by the ratio of this frequency to its cut-off frequency in the case of the low-pass filter or the inverse in the case of the high-pass filter. This ratio is usually known as the " $a$ " of the sections and $a$ is always greater than I. (We can, in fact, now describe the prototype section as being merely a special case of the derived section where $a=$ infinity.)

The attenuations obtainable with sections having various values of $a$ are shown in Fig. 13 where $f_{c}$ is the cut-off frequency. In these curves it is assumed that the coils used have a ratio of reactance to resistance (Q) of ioo which should be quite practical at

Fig. 12. Filter design formulae.

these frequencies. (The coils should be either wavewound or wound on dust corrs using formers with at least fourslots in order to keep down their capacitance.)

The characteristic imp dance of various sections -by which is meant the impadance looking into such a section when terminated with an infinite train of similar sections-will vary with frequency, and only equals the value $R$ at zero friquency in the case of the low-pass and at infinite frequency in the case of the high-pass sections. For all values of $a$ the derived sections quoted in Fig. 12 have the same characteristic impedance as the prototype sections, and therefore the two types can be connected together without introducing any impedance mismatch. But the mid-section impedance of a derived half-section, when correctly terminated with its characteristic impedance, will not be equal to the characteristic impedance of the whole section, but will vary with $a$ as is shown in Fig. I4 where the mid-section inpedance is expressed in relation to $R$. For certain valucs of $a$ around 1.25 the ratio is constant over a large portion of the pass band of the filter. Such a half-section is therefore very useful for terminating a filter, as its mid-section impedance is approximately a constant r sistance.

An inspection of the curves of Fig. I3 shows the relative inefficiency of the sections having low values of " $a$," for not only do they put in less attenuation at their resonant frequency but they require to be used in conjunction with other sections of higher value of " $a$ " if the filter is to maintain its attenuation in the stop range. For instance, it becomes necessary to follow a section having $a=1.02$ by a section of $a=1.09$, if the attenuation is to be maintained beyond the cut-off frequency. Similarly, a section of $a=1.09$ will require to be followed by a section having an " $a$ " of about 1.6 and so on. There are, of course, an infinite number of values of " $a$ " which might be chosen, but as it is impossible to give a complete set of filter tables in an article such as this, the above values for " $a$ " have been chosen as being representative. These particular values are advantageous because $a=1.02$ gives $m=0.2, a=1.09$ gives $m=0.4$, while $a=1.6$ gives $m=0.8$. As these three values of " $m$ " are each double the previous value, it will be seen from the formulae given in Fig. 12 that using such values there will be a great deal of similarity in the component values of the various sections. Similarly $a=$ 1.06 gives $m=0.33, a \Rightarrow$ 1.35 gives $m=0.66$ and $a=$ infinity gives $m=1$, and again using these sections together we may expect to get a certain similarity in component values.

A further inspection of the formulae in Fig. 12 will reveal that in the formulae for the prototype low-pass filter

$$
\mathrm{L}=\frac{\mathrm{R}}{\pi f_{c}} \text { and } \mathrm{C}=\frac{\mathrm{I}}{\pi f_{c} \mathrm{R}}
$$

Eliminating R we get,

$$
\begin{equation*}
\mathrm{LC}=\frac{\mathrm{I}}{\left(\pi f_{\mathbf{c}}\right)^{2}} \text { or } f_{c}=\frac{\sqrt{\mathrm{LC}}}{\pi} \tag{I}
\end{equation*}
$$

From this we see that we can vary the cut-off frequency by altering either L or C or both.

Eliminating $f_{c}$ we get,

$$
\begin{equation*}
\frac{\mathrm{L}}{\mathrm{C}}=\mathrm{R}^{2} \tag{2}
\end{equation*}
$$

From equations ( 1 ) and (2) we see that if we alter both L and C maintaining the ratio $\frac{\mathrm{L}}{\mathrm{C}}$ constant, we can


Fig. 13. Attenuation curves of typical filter sections.
alter the cut-off frequency of the filter and at the same time keep R constant, and this is, in fact, what is generally done.

If, however, we are prepared to let the value of $R$ vary according to equation (2), then by equation ( 1 ) we can vary the cut-off frequency by altering either C or L . In practice it is easier to alter C , as ganged condensers are readily available commercially.

Considering the medium-wave band, the frequency range is from about $600 \mathrm{kc} / \mathrm{s}$ to $1,500 \mathrm{kc} / \mathrm{s}$, or a ratio of 2.5. By equation ( I ) this will require a change of L or C by $2.5^{2}$ and this in turn will cause $R$ to alter by a ratio of 2.5 . To be absolutely correct we could make the termination of the filter a variable resistance and adjust it to the right value for the particular cut-off frequency. In practice this is not necessary, and if it is made a fixed resistance of $\sqrt{2.5}$ its correct value when the filter is adjusted for a cut-off of $600 \mathrm{kc} / \mathrm{s}$, then the error introduced is not serious and we need not worry about it.

If it is chosen to alter the cut-off frequency by varying $C$ (keeping $L$ constant), then the " $\pi$ " type

Fig. 14. Mid-section impedance of derived sections expressed relative to $R$, plotted against frequency relative to $f_{c}$.


of low-pass filter and the " $T$ " type of high-pass filter should be avoided, as the derived forms of these sections call for two variable condensers of different value, which are not readily available in ganged form.

By way of an example, let us consider the design of a high-frequency filter with a cut-off ratio of r.og. Using the " $T$ " form of the low-pass filter, we can keep all the condensers to the same value if all the sections are derived and the half sections at the end have an " $m$ " of double the middle sections, i.e., if the middle sections are $m=0.4(a=1.09)$ and the half sections at the end are $m=0.8(a=1.6)$, then all the condensers in the filter are 0.4 of the value of the condenser in the prototype section, which is

$$
\begin{equation*}
\mathrm{C}_{l p}=\frac{0.4}{\pi f_{l p e} \mathrm{R}} \tag{3}
\end{equation*}
$$

where $f_{l p c}$ is the cut-off frequency of the low-pass filter.

Again, using the " $\pi$ " section of the high-pass filter, we can keep all the condensers in the filter to the same value if all the sections are derived and if the
half sections at the end have an " $m$ " of double the middle sections. Again, using $m=0.4$ ( $a=1.09$ ) for the middle sections and $m=0.8(a=1.6)$ for the half sections at the end, the condensers are $\frac{1}{0.4}$ of the condenser for the prototype section, which is

$$
\begin{align*}
C_{h p} & =\frac{1}{4 \pi f_{h p c} R} \times \frac{\mathrm{I}}{0.4} \\
& =\frac{1}{1.6 \pi f_{h p \mathrm{c}} R} \tag{4}
\end{align*}
$$

where $f_{h p c}$ is the cut-off frequency of the half-pass filter.

Bearing in mind (Fig. 3 of previous instalment) that we required $f_{h p c}$ to be 0.8 of $f_{l p e}$, we get

$$
\begin{align*}
\mathrm{C}_{h p} & =\frac{\mathrm{I}}{\mathrm{I} .6 \pi \mathrm{o} .8 f_{l p c} \mathrm{R}} \\
& =\frac{0.8}{\pi f_{l p c} \mathrm{R}} \text { (approx.) } \tag{5}
\end{align*}
$$

From the above it will be seen that we have now got all the condensers in the low-pass filter the same and all the condensers in the high-pass filter equal to and double the value of those in the low-pass filter. This was another important reason in the choice of these particular values of " $a$ " as it means that standard condensers can be used throughout, the only difference between high- and low-pass filters being that each condenser in the former is built by putting two condensers in parallel.

The above completes the design considerations for the high-and low-pass filters except to give an idea of practical values. If a commercial $0.0005 \mu \mathrm{~F}$ ganged condenser is made the basis of the design, then putting $\mathrm{C}_{l y}=0.0005 \mu \mathrm{~F}$ when $f_{l p c} \stackrel{c}{=} 600 \mathrm{kc} / \mathrm{s}$ in equation (3), we get $R=$ 400 ohms at $600 \mathrm{kc} / \mathrm{s}$.

As already explained, a


Fig. 19. Overall a.f. response of the receiver for various positions of the variable i.f. filter.
suitable mean terminating resistance would be this value multiplied by $\sqrt{2.5}$, i.e., 650 ohms.

The circuit of the above filter is given in Fig. 15 , its calculated characteristic is give I in Fig. I6 and its measured characteristic at various points on the mediumwave band is given in Fig. 17.

The above is a typical r.f. filter. The i.f. filter follows on similar lines but from consideration of component values and in order to work at a suitable intervalve impedance, it is designed with a characteristic impedance of I 2,000 ohms. In addition the con-

Fig. 18. Measured characteristics of variable inter-mediate-frequency filter for various capacitance settings.


Fig. 20. Practical layout of a filter-tuned receiver as seen from the back.

Fig. 21. The three main assemblies. (Left) the i.f. filter, showing the ten fixed capacitors (air-spaced trimmers are used for convenience) and the eight variable condensers (four "split stator" condensers were used for convenience) connected by gears. (Centre) the screened coil assembly and (right) the r.f. filter showing the wire and drum drive.
densers are not made completely variable, but consist of a fixed portion and a io per cent variable portion : further the variable portions in the high- and low-pass sections of the filter are connected in opposition in such a way that the cut-off frequency of the one rises as the other falls and finally the ratio of the high- and low-pass variable condensers is chosen to be such that for any movement of the condensers both filters change their cut-off frequency by the same amount in cycles per second

## Overall A.F. Response

The ineasured characteristic of this filter for various positions of the condenser shaft is shown in Fig. I8. (The dotted curve represents the characteristic of a sharply-tuned circuit which is in the " magic eye" circuit and enables the receiver to be tuned to the centre of its i.f. characteristic-an important requirement in such a receiver as this.) The audio-frequency response of the receiver for various positions of the i.f. condensers is shown in Fig. 19.

In its simplest form the practical design of such a receiver is shown in Fig. 20 which is a back view. It contains five sub-chassis namely :-
(I) The decoupling resistor and condenser assembly.
(2) The valveholder assembly.
(3) The r.f. filter assembly.
(4) The screened coils assembly.
(5) The i.f. filter assembly.

The last three assemblies are shown removed in Fig. 21 which indicates the method of ganging the condensers in the r.f. and i.f. filters. In the former case a wire and drum drive is used, and as two of the condensers are above earth potential they are mounted on a sheet of Paxolin and the drums are made of ebonite: no difficulty is experienced in ganging the condensers if the grub screws in the drums are tightened when all the condensers are maximum. The condensers are not likely to get out of gang if suitable wire is used, but if they should they can quickly be re-ganged by slackening off the grub screws, setting all the condensers to maximum and then retightening the screws.
In the case of the i.f. filter, air dielectric trimmers have been used for the fixed portion of the capacitances and the variable portions are coupled together by means of gears, though a wire and drum drive could equally well have been used.

Space has precluded giving complete details of the receiver design but it is hoped that enough has been given to show the general lines of approach. The resulting receiver has much to recommend it, particularly its relative freedom from noise and its variable i.f. filter, which can be adjusted continuously and without introducing clicks or noise of any kind until the best position is found. It is,

however, a complicated design and cannot be built without adequate test equipment. Furthermore, when built it is still only a compromise solution. In the author's opinion, it is the best compromise possible in modern listening conditions.

## MOUSE ©F ${ }^{\text {COMMONS S.R.E. }}$

APART from the general interest in the sound reinforcing system in the new Chamber of the House of Commons this low-intensity installation is of particular interest in that it has been designed to ensure that all members can hear and, if desired, be heard without having to leave their seats.

To achieve this a large number of low-intensity loudspeakers have been installed in the woodwork at the back of the seats-one unit for two members-and six uni-directional microphones have been suspended from the ceiling over the front benches and three each side, mounted on cantilever arms, over the back

Loudspeakers behind bronze grilles in the carved oak rail above members' seating in the new House of Commons.

benches. The selection of the microphones during debates and the general supervision of the installation is undertaken by an operator in a cubicle overlooking the chamber.

To obviate "howl" caused by acoustic feedback when microphones and loudspeakers are in close proximity, the reproducers are arranged in groups which are co-related with the pick-up area of the microphones. When a microphone is brought into circuit the reproducers in the immediate vicinity are muted and in certain cases the input to those in adjacent areas is attenuated.

The main equipment comprises fourteen microphone pre-ampifiers with stand-by units. After preamplification, frequency and volume correction, the signals are passed through a "buffer" amplifier to raise the level sufficiently to energize the power amplifiers. There are eighteen of these, plus spares, and each one provides power for one loudspeaker zone, the input to which is connected through two relays and attentuators to provide, automatically, the correct degree of muting or volume reduction associated with the microphone employed in each particular zone. Separate amplifiers are provided for operating db volume indicator meters, which provide visual indication of output signals at the main operating position and at the alternative control position below the floor of the House. A periscope is provided at this position to enable the operator to view the chamber and select the appropriate microphone.
In addition to the microphones and loudspeakers already referred to there is a microphone above the Speaker's chair and another on the Table of the House. It should be pointed out that the reproducers on the tables in the Press Gallery and in other galleries are operated at a constant level.
The entire installation was undertaken by Tannoy Products (Sound Rentals, Ltd.).

# International Vision Standards - Short-Wave Broadcasting * R.T.E.B. Exam. Results - Training Opportunities * Television Relay Service 

## European Television

DETAILS of the standards drawn up by the Television Study Group of the International Radio Consultative Committee (C.C.I.R.), following the investigation of systems at present in use in France, the U.S. and the U.K., have been published by the European Broadcasting Union, and we give below a summary of them:-

The width of the channel is to be $7 \mathrm{Mc} / \mathrm{s}$ (U.S. standard is $6 \mathrm{Mc} / \mathrm{s}$ and the British $5 \mathrm{Mc} / \mathrm{s}$ ) with the vision carrier $5.5 \mathrm{Mc} / \mathrm{s}$ below the sound carrier-in this country, of course, it is $3.5 \mathrm{Mc} / \mathrm{s}$ below.

There are to be 625 lines per picture with interlaced scanning. The line and frame recurrence frequen cies are $15,625 \mathrm{c} / \mathrm{s}$ and $50 \mathrm{c} / \mathrm{s}$, respectively. The British line frequency is $10,125 \mathrm{c} / \mathrm{s}$. The aspect ratio of the picture is that now used in this country-4:3, horizontal to vertical.

The lower sideband of the vision channel is to be partially suppressed, and, as in the U.S., negative modulation will be used. The pedestal level (normal black level) will be represented by a definite carrier level independent of light and shade in the picture, and this will be transmitted at 75 per cent-with a tolerance of $\pm 2 \frac{1}{2}$ per cent-of the peak carricr amplitude. Peak white will be at least to per cent of the full carrier amplitude.

It is recommended that frequency modulation be used for the sound transmission. No reference is made in the recommendations to the polarization of the transmissions.

## H.F. Broadcasting

W
HEN the Hague Radio Conference, due to have opened in September, was called off (see page 318, September issue) the delegates to the International HighFrequency Broadcasting Conference, which had been in session, first at Florence and then Rapallo, since April 1st, decided by 39 votes to 13 (with 4 abstentions) to conclucle their deliberations. This was on August igth. The main reason for discontinuing all work on pre paring a plan for h.f. broadcasting was that such a plan, based on the new bands allocated at Atlantic City (1947), could be implemented only if another Conference, such as that planned for the Hague, made available other frequencies to the
services at present occupying these bands.

The Conference was convened to draw up plans for the allocation of frequencies (between 5.95 and $26.1 \mathrm{Mc} / \mathrm{s}$ ) to the world's shortwave broadcasting stations on the basis of time sharing and simultaneous channel sharing. These plans were hased on the various phases of the sun-spot cycle

Being unable to produce the plan envisaged, the Conference, at its final meeting, made a number of recommendations which are in cluded in the September i5th issue of the European Broadcasting Union Bulletin. To encourage the use of the higher frequencies for broadcasting and thereby reduce the congestion in other sections of the h.f. band, it is recommended that the frequency coverage of receivers should be extended to include the upper limits of the band

## Servicing Examinations

THE result of the first Television Servicing Certificate examination held last May by the Radio Trades Examination Board, which was restricted to the London area, shows that of 30 entries 16 passed and 12 were referred in the pracrical examination. Of the 264 candidates who entered for the Radio Servicing Certificate examination, also held in May, I 37 were successful and 45 were referred in the practical tests.

Application forms for the next examinations, to be held in May, 1951, may be obtained from the R.T.E.B., 9, Bedford Square, London, W.C.I. The closing date for entries for the Television Certificate is January 15 th and that for the Radio Certificate is February 1st.

## Frequency Control

THE need for a simple method of quoting frequency staliility was stressed by C. F. Booth, the new chairman of the I.E.E. Radio Section, in his inaugural address on " The Evolution of Frequency Control." A proposed methot, that he felt was worthy of consideration, is to express frequency stability in terms of the common logarithm of the reciprocal of the frequency variation when the latter is expressed as a fraction; for example, an oscillator subject to frequency variations of 1 in $10^{6}$ would have a stability of 6 stability units, or, say, 6 S.U.

## Technical Experience Abroad

$\mathrm{O}^{\mathrm{F}}$the 1,672 University and College students who, through the International Association for the Exchange of Students for Technical Experience, gained experience in industry overseas during the summer vacation this year, 353 were from Great Britain, 20 from France, 197 from the Netherlands and 184 from Sweden. Twelve countries, inclucd ing, for the first time, the U.S.A and Germany, took part in the exchange scheme. Great Britain and Sweden headed the list of countries receiving students with an intake of 368 and 344 respectively
The third Annual Report of the Association records that there was - an increase of 436 on the exchanges made in 1949. About a dozen British radio firms received overseas students under the scheme.
Particulars of the Association, which, it is stressed, organizes these exchanges on a reciprocal basis, are obtainable from the General Secretary, J. Newby, Imperial College South Kensington, London, S.W. 7

## B.B.C. Changes

FOLLOWING the much-talked-of resignation of Norman Collins as Controller of B.B.C. Television, the politics of which we will not enter into, George Barnes, who has been B.B.C. Director of the Spoken Word since the creation of the post in 1948, has been appointed to the new post of Director of Television. Norman Collins has stated that he has left the Corporation because of the principle which was being adopted in the development of television-its merger "into the Colossus of sound broadcasting."
T. W. Chalmers, Controller of the Light Programme, has been seconded for three years to become head of the new Nigerian Broadcasting Service (see "In Brief") and has been succeeded by Kenneth Adam.

## Electronics Scholarships

$\mathrm{T}^{0}$meet the growing need for technicians in the research and design sections of E.M.I. a scholarship scheme for a special four-year course in electronics has been prorided by E.M.I. Institutes.
The scheme provides for a grant of $£ 50$ p.a. to each successful applicant towards the $£ 400$ course and a
maintenance grant of at least $£ 50$ p.a. Candidates, who must be between 16 and 18 years of age, and preferably of Higher School Certifcate standard in science, must enter the company's service for four years after the satisfactory completion of the course.
Application forms and particulars of the course, which provides for three years at the Institute and one year's practical experience in the company's factories and workshops, are obtainable from the Principal, E.M.I. Institutes, io, Pembridge Square, London, W.2. The first course commences on January 17 th.

It is stated that, in normal circumstances, students will receive a deferment from National Service during the period of the course.

## News in Morse

SINCE the publication of the schedule of the morse transmissions of the London Press Service in our May issue, a number of changes have been introduced, and we give below a revised list. In addition to the morse transmissions listed, the speed of which is between 20 and 27 w.p.m., a large number of bulletins are transmitted by Mellschreiber.

| G.M.T. |  | Call | Freq. ( $\mathrm{Mc} / \mathrm{s}$ ) | Areas |
| :---: | :---: | :---: | :---: | :---: |
| 0030-0130§ | ,.. | MIK | 9.725 | 1 |
|  |  | GPJ | 10.885 | 2 |
| 0045-0230* |  | MIK | 9.725 | 1 |
|  |  | GPJ | 10.885 | 2 |
| 0130-0230§ | .... | MIK | 9.725 | 1 |
|  |  | GPJ | 10.885 | 2 |
| 0130-0300* | $\ldots$ | GDI | 7.780 | 3 |
|  |  | GAH | 8.065 | 4 |
| 0130-0300 |  | GCX | 8.920 | 5 |
| 0945-1045* |  | GCV | 19.365 | 6, 8. |
| 1100-1200\% |  | GCV | 19.365 | 6, 8. |
| 1115-1215\|| |  | GAG | 17.105 | 7 |
| 1200-1300\\| | ... | GCF | 19.005 | 5 |
| 1215-1315 $\ddagger$ |  | GAG | 17.105 | 7 |
| 1600-1700* |  | GCF | 19.005 | 5 |
|  |  | GIBI | 10.865 | 6, 8. |
| 1600-1800* |  | GIB | 11.980 |  |
| 1700-1800\|| | $\ldots$ | GBI | 10.885 | 6 |
| 1815-1930 | ... | GBG | 9.395 |  |
|  |  | GBI | 10.865 | $6,8$. |
| 1845-1945* | ".. | GIM | 12.957 |  |
|  |  | GBO | 13.665 |  |
| 1945-2215* | $\cdots$ | GDT | 9.925 | 6, 8. |
| 2045-2200 $\dagger$ |  | GBG | 9.395 9.395 | 7 |
|  |  | GDT | 8.925 | 6, 8. |
| 2100-2200* | $\ldots$ | GBO | 13.665 | 5 |
| 2251-00330* |  | GPX | 11.645 | 1 |
|  |  | GPJ | 10.885 | 2 |
| 2330-0030 $\dagger$ | $\ldots$ | GIB | 11.980 | 5 |

[^2]The number in the fourth column of the table denotes the area for which the transmissions are destined: I, N. America; 2, S. America; 3, Distant Europe; 4. Middle East; 5, Africa; 6, N.E. Asia; 7, S.E. Asia; 8, Australasia.

There is no restriction on the reception of these transmissions, which are radiated by Post Office stations, and the use of their contents outside the United Kingdom.

## National Exhibition

THE Radio Industry Council has announced the date of next year's National Radio Show-August 28 th to September 8th. As already stated, the show-the 18 th in the series-will be at Earls Court, London, and not Olympia, so that the well-established title Radiolympia will not be applicable.

## Greenwich Time Signal

AMEAN time signal is now being transmitted twice daily from the Royal Greenwich Observatory, Abinger Common, Surrey, in addition to the present rhythmic signal. The schedule is:-
> e.m.T.
> $\begin{array}{ccc}\mathrm{h} & \mathrm{m} & \mathrm{s} \\ 09 & 54 & 00\end{array}$

> to | 09 | 54 | 00 |
| :---: | :---: | :---: |

Preamble: "GBR GBR TIME" ( 4 times) followed by tuning dash.
095500
to 100000
Mean timg signal: Seconds dots, lengthened to dashes at the minutes.
100100
$0 \quad 100600$
o 100600
Rhythmic signal: Dots spaced 61 to the minute, lengthened to dashes at the minutes.
100605
0100615
Tuning dash.
This sequence is repeated from ${ }^{1754}{ }^{1}$ The
The morning transmission is radiated by GBR, $16 \mathrm{kc} / \mathrm{s}$; GIC, $8,640 \mathrm{kc} / \mathrm{s}$ and GIA, $19,640 \mathrm{kc} / \mathrm{s}$, and the evening transmission by GBR, $16 \mathrm{kc} / \mathrm{s}$; (iIC, $8,640 \mathrm{kc} / \mathrm{s}$ and GKU3, $12,455 \mathrm{kc} / \mathrm{s}$.

The duration of each dot is approx. o.r sec, and that of the dashes at the minutes is approx. 0.6 sec . The beginning of each dot or dash is the timing reference point.

## Hotel Television

A TELEVISION relay service has been installed in each of the 940 bedrooms in the Cumberland Hotel, London, W.r. The hotel was already equipped with an a.f. distribution system giving a choice of four programmes, and the same witing is used for television.

A four-element aerial is employed on the roof of the building to minimize the pick-up of interference and its output is fed to seven amplifiers, of which one is a spare. With a maximum gain of 55 db they provide an output of 2 V r.m.s. in 75 -ohm feeders.

The audio wiring of the hotel is carried out in 63 vertical ducts each feeding about 16 rooms. The r.f. television signal is fed by 75 -ohm balanced cable to the tops of these ducts, each amplifier feeding $9-12$ ducts. They are there connected by resistance hold-off pads to the audio wires which are used in a quad phantom circuit. A $\lambda / 4$ stub is connected a quarter-wavelength be yond the junction to prevent r.f.s. passing back to the a.f. source.

Resistive stopper pads are in-
cluded at the loudspeaker for the audio channel to divert the television r.f. signal to the outlet socket. Standard television receivers are used and the hotel charges 3 s a night for the service.

The equipment was installed by British Relay Wireless

## Television Components

$I^{N}$N preparation for the expected demand for television receivers when the projected stations open in Montreal and Toronto, the Canadian Government has authorized the release of U.S. dollars to fifteen Canadian set manufacturers for the purpose of buying components in the U.S.A. Attention having subsequently been drawn to the availability of television components in the U.K., the Canadian Department of Trade and Commerce has agreed, in principle. that where practicable the components should be secured in this country. British manufacturers who are represented in the Dominion, and wish to bring their components to the notice of the Canadian set manufacturers, should send catalogues and price lists in duplicate to E. J. McWilliams, Assistant Director, Emergency Import Control Branch, Department of Trade and Commerce, Ottawa. Firms without agents in Canada may like to use the services of the Board of Trade to secure representation, in which case communications should be sent to the Commercial Relations and Export Department, Thames House North, Millbank, London, S.W.i, quoting reference C.R.E. $5902 / 50$.

## IN BRIEF

Consol.-The second edition of the Ministry of Civil Aviation publication "Consol-A Radio Aid to Navigation" (M.C.A.P. 59), which has been published by H.M. Stationery Office. price is 6d, contains complete details of the service and coverage of the existing Consol stations at Bushmilis, Stavanger, Lugo and Seville and also of the service to be provided by the new station at Plonéis, near Quimper, Finistere (France), which will be brought into use in the near future.

Amateurs and Distress Calls.-In contrast with the facilities provided by American amateurs for a network of stations in cases of emergency, such as forest fires, floods, etc., which were tested during a recent nation-wide exercise, the Ministry of Transport has declined the offer made by the Radio Society of Great Britain for British amateurs to place themselves at the disposal of the appropriate authority for receiving distress messages fron vessels at sea.
M.F. Air Navigation.-It is pointed out by the Ministry of Civil Aviation that, while every effort is made to assign to m.f. navigational facilities trequencies from the aeronautical bands ( 255 to $285 \mathrm{kc} / \mathrm{s}$ and 315 to $405 \mathrm{kc} / \mathrm{s}$ ), it is not always possible to do so be-
cause of the congestion within those bands. Use is, therefore, made of broadcasting stations as non-directional beacons and of other non-directional beacons operating on frequencies outside the aeronautical bands but within the limits of 200 to $1,750 \mathrm{kc} / \mathrm{s}$.

Netherlands F.M.-According to information issued by the European Broadcasting Union, the Netherlands Government is planning the erection of a network of 12 f.m stations operating on frequencies between 92.5 and 94.7 $\mathrm{Mc} / \mathrm{s}$.

Suppressing the ignition system of motor wehicles is dealt with in one of the chapters of the fourth edition of "Automobile Electrical Equipinent," by A. T'. Young and L. Griffiths, which is issued by our Publishers for our associate journal Antomotrile Engineer. This 386 -page book, which deals exhaustively with electric lighting, starting and ignition as applied to the internal combustion engine, costs 25 s.

Multitone. The results of investigations, using the "Optimeter," in the volume requirements of five hundred hearingaid users are given in the article "How Loud Should I Listen?" included in the first issue of Hearing Aid News which will be issued at irregular intervals by the Multitune Electric Co.
Hearing Aids.-The National Institute for the Deaf bas issued a booklet, No. 48I, giving approved lists of manufacturers and suppliers of hearing aids in the United Kingdom. Makers of mechanical as weld as electronic aids are listed and those supplying group hearing aids are indicated.
Christmas lectures "'adapted to a juvenile auditory," which will again be given at the Royal Institution this year, will be on "Waves and Vibrations," by Prof. E. N. da C. Andrade, D.Sc., F.R.S. This 121 st raurse of six lectures, the fee for which is ins bid or children between 10 and 17 years of age, will be given on December 28 th and 30 th and January 2nd, 4 th ${ }_{k}$ 6th and gth at 3 .

"The Use of Radar at Sea" is the provisional title given to a textbook on the operational use of marine radar, which is in course of preparation by the Institute of Navigation. The book will contain contributions by twelve authors and will deal with every aspect of the subject. This is announced in the annual report of the Institute for the year ended June 30th, 1950, in which it is reported that the membership is now over I, Ioo.
"Decca News" is the title given to the recently introduced journal of the Decca Navigator Company, which at present is issued bi-monthly. Descriptive articles on both radar and Decca Navigator installations are included.

Post-graduate Training. - Twentyeight university graduates and technical college students recently commenced a special training course at the New Southgate Vorks of Standard Telephones and Cables. The course, which lasts a year and is the fourth since the conmencement of the company's present post-graduate training scheme, is designed to provide opportunities for candidates to relate their theoretical and practical training to the whole field of activities within the Standard organization.
Convention Cancelled. - The Radio Section of the I.E.E. had proposed to hold a convention covering "Current Researches in Electron Valves" during the Festival of Britain, but having been advised that to hold such a convention in prevailing circumstances might not be in the National interest, it has been decided not to proceed with the arrangements.

Interference Suppression.--The tele-vision-conscious motorist who fits Lucas suppressors to his car ignition system will be easily identified by a transparency bearing the words " This car has been suppressed," which is being fixed to windscreens. At both the recent Motor Shows in London the firm demonstrated the effects on television reception of suppressed and unsuppressed car ignition systems which, according to the Radio Industry Council, are responsible for 85 per cent of the interference caused. Lucas have recently produced a prototype of a distributor incorporating a carbon brush of resistive material which acts as a suppressor.
Nigeria, which at present relies on a wirel-witeless distribution system for radio programmes is to be served by six broadcasting stations to be erected at a cost of about £350,oon, which is being shared by the Nigerian Government and the UK.

COMET RADIO.-The Marconi radio installation in the de Havilland "Comet" jet air liner. The equipment shown includes intercom. units, direction finders (duplicate control units for which are in the cockpit) and high-power transmitters.

Instrument Show.-Scientific and industrial instruments are to have an exhibition of their own in London next year. Among the supporting organizations of the exhibition, which will be held at Olympia from July 4 th-14th, is the Scientific Instrument Manufacturers Association.

Television in Industry.-The use of television to permit the remote observation of dangerous industrial processes was recently demonstrated when the English Electric Co. employed it to show the operation of high-power switch gear. Standard $405-$ line Marconi apparatus was used.
Blind Technicians.-The Report of the National Institute for the Blind records that an experimental training scheme has been arranged with Electric and Musical lndustries to ascertain to what extent the blind can be instructed to undertake receiver repairs. This has necessitated the provision of circuit diagrams in relief and Braille-calibrated test equipment.
"Broadcast News," which is published by the Radio Corporation of America and is devoted almost entirely to descriptions of transmitting equipment for sound and vision, is now available to subscribers in this country from RCA Photophone, 36, Woodstock Grove, London, W.12. It is published at two-monthly intervals and the subscription rate is $t^{2}$ for twelve issues.

Vox Populi.-The International Short-lVave Cluh has announced the result of a census taken among its members in all parts of the world to ascertain the most popular short-wave broadcasting station. Leopoldville (OTC) comes first with 609 votes, Kadio Australia second with 446 and Switzerland third with 435, followed by Canada ( $\ddagger 19$ ), B.B.C. Overseas Service (401), Hilversum (388), The Voice of America (380) and many other stations with fewer votes

Sound Letters-magnetically recorded mescares on metallized paper tape-nase bean introduced by the Welfare Council of the Norwegian whaling fle t for use by the men working on the factory ship and their relatives. 1 hey are regularly flown tu and from the Antarctic whaling grounds.
B.I.F.-Next year's British Industries Fair will be held from 3oth April to ith May, at Olympia and Earls Court, London, and Castle Bromwich, Birmingham.
Institute of Physics announces an increase in its membership of some 200 during 1949.

Airborne Television.-A Marconi image-orthicon camera and 30-watt "suitcase" transmitter were used by the B.B.C. for the first television broadcast from the air which was radiated from the London and Sutton Coldfield stations on October Ist.

## BUSINESS NOTES

Rural Radio.-The Persian Government is considering the purchase of 50,000 radio receivers for use in rural areas. They should be batteryoperated, six-valve sets, fitted with "magic eye" tuning and covering the 60 to 120 -metre and 180 to 360 -metre bands. Interested firms should communicate with Esfandiar Bouzourg-
mehr, Director, Government Propaganda Department, Tehran, forwaiding details. and prices of their products. It is stated by the Board of Trade that it will be appreciated if copies of correspondence are sent to the Counsellor (Commercial) British Embassy, Tehran.
R.T.S.-We understand that it is only the stock and plant of the transformer business formerly operated by R.T.S. Electronics which has been acquired by EARL Services and to which reference was made in our last issue. R.T.S. are continuing the manufacture of the "Motoradio" car set and the "Minicomm" intercommunication equipment.
I.A.L.-Consequent on the withdrawal of the R.A.F. from Hargeisa Airfield, British Somaliland, International Aeradio, Ltd., will provide an Airport Controller for service in the Protectorate.
A contract for the supply of Decia navigational radar gear to be installed in ships of the Royal Navy has been received by Decca. The company states that orders to equip nearly five hundred naval and merchant ships have been received during the past twelve months.

British Radar for Japan.-The first Japanese ship to be fitted with British marine adar cquipment is a whale-oil refinery vessel which was equipped by Cossor Radar prior to leaving for the Antarctic.
Marconi Marine Radio gear, including two transmitters, two receivers, direction finder and echometer, is being installed in the new 28,000 -ton turbine tanker Verena under construction for the Anglo-Saxon Petroleum Company. The radio-communication and navigation equipment installed in the new Argentine liner 77 de Oclubre has also been supplied by Marconi's.

Trawler Radar.-Ten trawlers now under construction in this country for the Icelandic Government are to be equipped with Decca radar gear, Type r99.

Pakistan.-Seven more slips of the Pakistan merchant fleet are being fitted with radio equipment by the Marconi International Marine Communication Co. I.td.

Plessey announce the resignation of Wing Commander G. C. Cunningham, O.B.E., who was the company's communications sales manager.
Aerialite, Ltd.. of Stalybridge, Cheshire, announce that O. E. Trivett M.Brit.IRE., A.M.I.F.E., has been appointed personal assistant to the Managing Director. Before taking up this appointment he was concerned with the setting up of a plastics extrusion and radio factory in India in conjunction with General Electric of America.
Ersin Multicore Solder is being exported to an increasing number of countries; among the latest are Brazil, Argentina, South Africa, Syria and Mexico. In addition agents have recently been appointed in Iceland and Iran.

Oldham $8<$ Son, the battery manufacturers, are to establish a factory in Madras, India, which is to be managed by Oldham \& Son. India, Ltd.

Denco announce that, following the recent resignation of $D$. W. Heightman from the Board, A. W. Allwright has rejoined the company as Managing Director.
I.M.R.C.-The Brazilian Navy training ship Almirante Saldanha, recently refitted at Barrow, has been equipped with m.f. and h.f. transmitters by the International Marine Radio Company. The vessel was equipped by I.M.R.C. when built in 1934.

Sobell Model 610, six-valve superhet has been tested by the School Broadcasting Council for the United Kingdom and approved for use in schools
E.E.G. Exports.-Edison Swan have recently received orders from Spain, Poland, Egypt and Cevion for their Mark II electro-encephalograph.
E. K. Cole announce that Bentley Jones has resigned from the position of sales manager of the conipany's radio division. He has joined Thorn Electrical Industries as commercial director of the Ferguson television and radio division.
Dubilier Condenser Co. announce that F. H. McCrea, Managing Director, is undertaking an extensive businezs tour of Canada and the United States.
The Electrical Apparatus Co., of St Albans, announce that J. de Gruchy has been appointed Head of the Instru ment Department which at present is concentrating on the production of moving-iron and moving-coil meters.

## NEW ADDRESSES

British Standards Institution has opened an office at I2, Hilton Street Manchester, I (Tel. Central 4856). where a complete set of British Standards may be consulted and copirs purchased.

Marconi's have recently established an Inspectorate at Mombasa providing radio and radar maintenance for vessels calling at ports in the area.

Depots are being opened by Aerialite, Ltd., at 343. Deansgate, Manchester, 3 (Tel. BLA. 3524); and 50, Cathay, Red cliffe, Bristol (Tel. 26130).

Southern Radio Supply, Ltd., have moved from 46, Lisle Street, Lonton, W.C.I, to II, Little Newport Street, London, W.C. 2 (Tel. Gerrard 6653 ).

## MEETINGS

Institution of Electrical Engineers
Radic Sechion.-"Crustal Liodes," by R. W. Douglas, B.Sc., and E. G James, Ph. D., and "Crystal Triores," by T. R. Scott, B.Sc., at 5.30 on December 6th

Discussion on "Have Post-War Broadcast Receivers taken Full Advan tage of Wartime Development?" opened by R. B. Armstrong, B.Sc., at 5.30 on December I8th

The above mectings will be held at 5.30 at the I.E.E. Saroy Place, London, W.C.2.

Cambridge Radio Group.-"A Review of some Television Pick-up Tubes," by J. D. McGee, M.Sc., Ph.D., and "The Design of a Television Camera Channel for Use with the C.P.S. Emitron," by E. L. C. White, M.A., Ph.D., and M. G. Harker, B.Sc. (Eng.), at 6 on December 5th at the Cambridgeshire Technical College.

North-Eastem Radio Group--" Electronic Counters and Some Applicitions,' by R. B. Conn B.Sc., at 6.15 on December I8th at King's College, Newcastle-on-Tyne.
Sheffield Sub-Centre.-" Some Electromagnetic Problems," by Prof. G. W. O. Howe, D.Sc., LL.D., at 6.30 on December isth at the Grand Hot 1 l, Sheffield.
North Lancushive Sub-Centre.-" The General Aspects of Television," by A. J. Biggs, Ph.D., B.Sc., at 7 on December 20th at the Harris Institute, Corporation Street, Preston.
South Midland Centre-" Generation and Flow of Harmonics in Transmission Systems," by S. Whitehead, M.A., Ph.D., and W. G. Kadley', C.B.E., Ph.D.(Ens.), at 6 on December $4^{\text {th }}$ at the James Watt Memorial Institute, Great Charles Street, Birmingham.
Reading (Berks) Bistrict. -- Tele vision Engineering," by D. C. Birkinshaw, M.B.E., M.A., at 7 on Decemb IIth at the Great Western Hote? Reading.
British Institution of Radio Engineers
London Section. - Discussion on " Progress in Loudspeaker Design," opened by R. L. West, B.Sc., at 6.30 on December 15th at the London School of Hygiene and Tropical Medicine, Keppel Street, Gower Street, W.C.I.

North-Easlem Section.-"A Survey of Television Development and its Problems,' by H. J. Barton Cliapple, B.Sc., at 6 on December I3th at Neville Hall, Westgate Road, Newcastle-onTyne.

Scottish Section.-"Vacuum Engineering Applied to Electronics," by D. Latham, B.Sc., at 6.45 on December 7th at Heriot-Watt College, Edinburgh.
North Western Section.-'High Fidelity Reproduction," by H. J. Leak, at 6.45 on December 7 th at the College of Technology, Manchester.
British Sound Recording Association
" The Application of Magnetic Coatings to Film Stock," by G. F. Dutton, Ph.D., D.I.C., at 7 on December 6th at E.M.I. Studios, Led., 3, Abbey Road, St. John's Wood, ${ }^{\text {' }}$ London, N.V. $S$ (Joint Meeting with the British Kinematograph Society).
" Modern Recording Technique," by C. E. Watts at 7 on December rath at the Central Library, Guilihall. Portsmouth. Demonstrations from $5 \mathrm{p} . \mathrm{m}$. (Joint Meeting with the Institution of Electronics).
"The Crystal Pickup With Particular Reference to Long-Playing Records," by S. Kelly at 7 on Decernber 2oth at the Royal Society of Aris, John Adam Street, London, W.C. 2.

## Television Society

Engineering Group. - "Television Transmission for the Amateur" by M. Barlow (Hon. Sec. British Amateur Television Society), at 7.0 on December rath, at the Cinema Exhibitors' Association, 164, Shaftesbury Avenue, W.C. 2.

## Institution of Electronics

Midlands Branch.-' Electronic Instruments for Temperature Measurements and Control, and for Safeguarding Combustion," by H. A. Stevenson and C. Smith at 7 on December 5th at the Warwick Room, Imperial Hotel. Temple Street, Birmingham, 2.

# Tolerances and Errors 

"CATHODE RAY" Explains How They Add Up in the Final Result

SOMEONE has written to me to ask: what about tolerances? And to make his point clear he proceeded to fire a number of diverse examples in rapid succession. Some of these concerned combinations of resistors, and he ended by asking "Is there a 'tolerance arithmetic' for simple series/ parallel networks?" To tell the truth, I hadn't given the matter much thought, but after having done so I can answer with a definite " yes."

The subject of errors, even if confined to the field of measurement, is a very large one, and to the mathematically-minded a very attractive one. Large and excessively learned books have been written on it. But they concentrate on the question of probable error, which must be clearly distinguished from maximum error or, as it is called in engineering, tolerance. The tolerance is so important in connection with radio components that the colour code applicable to them has one unit devoted to indicating it.

So when one buvs even such a simple thing as a resistor one must be prepared to answer the question "Yes, sir; what tolerance?"-or look a fool. It is not merely an academic question; it has more to do with the price apparently, than anything else The reason lies in the fact that the manulacture of resistors (of the ordinary composition type) is rather like shooting at a target from a considerable distance. The resistances no more come out all exactly the right value than all the bullets go through a single hole exactly in the centre. In the absence of any special trends, the distribution of bullets relative to the centre of the target, and of resistances relative to the nominal value, is found to follow what is referred to in the learned books as a normal (or Gaussian) error curve, Fig. I. Over a narrow range on each side of the nominal value the quantities are distributed almost evenly. Those falling within $5 \%$ of the nominal value are picked out and marked " $\pm 5 \%$." From the remainder, those falling within $10 \%$ are picked out and marked " $\pm 10 \%$ " Similarly for the $\pm 20 \%$. Those not within $20 \%$ of one nominal value come within $20 \%$ of another nominal value and can be marked accordingly. So, on this ple:1, buyers of $\pm 10 \%$ resistors are unlikely to find them within $5 \%$, and the usual assumptions made by the "probable-error" school of thought fall down.

If all were offered at the same price, the makers would have no sale for the $10 \%$ and $20 \%$ stocks, so the prices are graduated to equalize the demand.

Generally one does not buy a resistor as a selfcontained unit, to be preserved for ever in complete isolation. It is intendéd to be used in conjunction with other components, to achieve some desired result. Even if it is wanted as a single fixed standard of resistance, the ultimately important matter is not the maximum error in that resistance but the error in the results of measurements based on it. So one ought to have a clear idea of how tolerances in components affect the maximum possible error in the final result. It is because I suspected that one cloes
not have this clear idea that the subject struck me as being just the thing for considering here.

The sort of question one ought to be able to answer can be indicated by an example: We want to pass a current of $50 \mathrm{~mA} \pm 7 \frac{1}{2} \%$ from a $200-\mathrm{V}$ supply and we have a $6.8 \mathrm{k} \Omega$ resistor with a tolerance of $5 \%$. This resistance is too high, because what is wanted is $4 \mathrm{k} \Omega$; but a $10 \mathrm{k} \Omega$ in parallel would do the trick (within about $1 \%$ ); what tolerance can this rok $\Omega$ be allowed?

## Series Combinations

Let us, then, tackle resistance networks, or at any rate those whose overall resistance can be calculated by the two simple principles of series and parallel addition. Before starting there are one or two things we ought to be quite clear about.

When a resistor is marked " $200 \Omega \pm 10 \%$ " (read as " 200 ohms plus-or-minus so per cent") it means that if its actual resistance is anything between $180 \Omega$ and $220 \Omega$ you have to be satisfied; but if it is outside those limits you can claim your money back. Dimensions in engineering drawings are sometimes marked with unequal positive and negative tolerances; but we shall exclude them, or (if you insist) alter the main dimension so as to divide the tolerance equally. We make this assumption so that we can simplify our calculations by using + signs for all tolerances, except where a positive overall error is contributed to by a negative tolerance, as for example in the effect of shunt resistance on meter current.
If we had been interested in probable errors we would have had to take account of the fact that when two quantities are combined, each having random variations from the nominal value, there is a lucky chance that the errors will tend to cancel one another out, as well as the unlucky chance that they may add up. So the most probable error is less than the sum of the two. But here we are taking a sternly engineering attitude, leaving nothing to chance, and (whatever we may secretly hope) being prepared for the worst. So we shall assume in every case that the tolerances combine in the worst possible way

If in an unguarded moment one were to be asked


Fig. I. If the components in a very large batch are arranged according to their departure from the nominal value, their quantities tend (theoretically) to follow a curve something like this.
about the overall tolerance when two equal $\pm 5 \%$ resistances are connected in series one might (without thinking) answer " $\pm 10 \%$ " But this would be confusing percentages (or proportional errors) with absolute errors. If the two resistors were each $1,000 \Omega$, each would have a tolerance of $50 \Omega$, so the tolerance in the combination would be $100 \Omega$, it is true. But relative to the resistance of the combination, $2,000 \Omega$, this would still be $5 \%$. One can go further and say that in any combination of resistances having equal percentage tolerance; the overall percentage tolerance is the same as that of the components. The proof, if one is wanted, will emerge as a special case of the less obvious pioblem of any-value tolerances.
This problem can be solved by at least three methods: by common sense, by simple algebra, and by diff rential calculus. I am leaving out of a.ccount a fourth method-by arithmetic-because that solves only one example at a time, and I assume that anybody reading this article is more interested in getting to the root of the matter.

The common-sense method is to argue that in a combination of any number of resistors in series the tolerances in ohms add up together, just like the nominal resistances, but when they are expressed as percentages of the whole combination they aff ct it only in proportion to the individual resistances. For example, if a $20 \%$ resistor comprises one quarter of the resistance of the combination, its contribution to the overall tolerance will be one quarter of $20 \%$ Similarly for the other resistors.

Simple algebra reaches the same result by denoting the resistances by symbols and subjecting them to the ordinary arithmetical operations. Suppose there are two resistances, nominally $\mathrm{R}_{1}$ and $\mathrm{R}_{2}$ (Fig. 2), and their respective tolerances, expressed in the same units as $\mathrm{R}_{1}$ and $\mathrm{R}_{2}$, are $r_{1}$ and $r_{2}$. The corresponding percentage tolerances, $p_{1}$ and $p_{2}$, are therefore $100 r_{1} / R_{1}$ and roo $\gamma_{2} / R_{2}$. The nominal value of the combination, which we shall denote by $R$, is of course, $R_{1}+R_{2}$; and according to the rules we have just laid down the tolerance of the combination (call it $r$ ) is $r_{1}+r_{2}$. What we want to find is $p$, the percentage tolerance of the combination. It is, of course

$$
\begin{aligned}
p & =\frac{100 r}{\mathrm{R}} \\
& =\frac{100\left(r_{1}+r_{2}\right)}{R_{1}+R_{2}} \\
\text { Since } p_{1} & =\operatorname{Ioor_{1}/R_{1},r_{1}\text {mustbe}p_{1}R_{1}/\text {Ioo.So}} \\
p & =\frac{\operatorname{IOO}\left(\frac{p_{1} R_{1}}{100}+\frac{p_{2} R_{2}}{100}\right)}{R_{1}+R_{2}} \\
& =\frac{p_{1} R_{1}+p_{2} R_{2}}{R_{1}+R_{2}}
\end{aligned}
$$

If we had had three resistances, it would have been

$$
p=\frac{p_{1} \mathrm{R}_{1}+p_{2} \mathrm{R}_{2}+p_{3} \mathrm{R}_{3}}{\mathrm{R}_{1}+\mathrm{R}_{2}+\mathrm{R}_{3}}
$$

and so on.
This just puts in symbols what we have already said in words. Perhaps this would be made clearer if we rewrote the formula thus:

$$
p=\frac{p_{1} \mathrm{R}_{1}}{\mathrm{R}}+\frac{p_{2} \mathrm{R}_{2}}{\mathrm{R}}
$$

etc. Example : A $200-\Omega \pm 10 \%$ resistor is connected in series with a $500-\Omega \pm 5 \%$ resistor; what is the overall percentage tolerance and the contribution

Fig. 2. Two resistances in series, the nominal values being indicated by capital letters and the tolerances by small.

of each resistor ? The first contributes $10 \times 200 / 700$ $=2.86 \%$, and the second $5 \times 500 / 700=3.57 \%$; total $6.43 \%$.

As a matter of fact it is rather silly to give this answer to two decimal places, because the resistor sorters certainly do not work to that degree of precision. Another reason is that in general the accepted methods of calculating tolerances are only approximate. As it happens, however, this simple case of adding two quantities is an exception, thanks to the custom of reckoning tolerances relative to the nominal values. You can check this by taking an extreme case : suppose the tolerance of the $200-\Omega$ had been $\mathrm{I} \%$ and that of the $500-\Omega$ had bsen $60 \%$. Then with positive errors the components would have been $202 \Omega$ and $800 \Omega$ respectively, so $R_{2}$ would have been nearly four times $R_{1}$. With negative errors they would have been $198 \Omega$ and $200 \Omega$, making $\mathrm{R}_{2}$ and $\mathrm{R}_{1}$ almost equal. So it would seem as if the proportions in which each error contributed to the overall result would be greatly influenced by the errors themselves. And so they would, if they were reckoned on the actual values, of if the errors interacted on one another (as would be indicated if there were terms containing both $p_{1}$ and $p_{2}$ ). Check it with the example if you like; the maximum error (positive or negative) is $302 \Omega$, which relative to the nominal value of the combination ( $700 \Omega$ ) is $43.15 \%$. The calculated contributions are $\mathrm{I} \times$ $200 / 700=0.29 \%$ and $60 \times 500 / 700=42.86 \%$, which add up to $43.15 \%$.

But what about the diff rential calculus method? In this case it offors no advantage whatever; in fact, it only tells us what we have already accepted as obvious, namely that $r=r_{1}+\gamma_{9}$. But it is quite otherwise with more complicated cases, which by simple algebra are usually unbearably tedious.

## And Now in Parallel

Even such an apparently simple situation as two resistances in parallel is bad enough. The nominal value $R$, according to the well-known formula, is $\frac{R_{1} R_{2}}{R_{1}+R_{2}}$.

We can see that a positive error in either leads to a positive error in the combination, so + signs throughout are in order, and the maximum value of the combination is

$$
\mathrm{R}+r=\frac{\left(\mathrm{R}_{1}+r_{1}\right)\left(\mathrm{R}_{2}+r_{2}\right)}{\left(\mathrm{R}_{1}+r_{1}\right)+\left(\mathrm{R}_{2}+r_{2}\right)}
$$

and $r$ is obtained by subtracting R from this. As before, $p=100 \mathrm{r} / \mathrm{R}$. Now you can go ahead and work it out, on the same principle as for resistances in series. But make sure you have plenty of paper ! The result, in spite of a considerable amount of boiling down and tidying up, is the rather unpleasant

$$
p=\frac{p_{1} \mathrm{R}_{2}\left(\mathrm{I}+\frac{p_{2}}{100}\right)+p_{2} \mathrm{R}_{1}\left(\mathrm{I}+\frac{p_{1}}{100}\right)}{\mathrm{R}_{1}\left(\mathrm{I}+\frac{p_{1}}{100}\right)+\mathrm{R}_{2}\left(\mathrm{I}+\frac{p_{2}}{100}\right)}
$$

But provided that the tolerance percentages are
reasonably small, one can get a sufficiently good approximation by neglecting them where they are divided by 100 ; thus:

$$
p \simeq \frac{p_{1} R_{2}+p_{2} R_{1}}{R_{1}+R_{2}}
$$

This is rather interesting, bocause it is the same as for resistances in serifs, except that each tolerance contributes in proportion to the other resistance. Which fits in with common sense, because if $100 \Omega$ is shunted by $10,000 \Omega$ the $100-\Omega$ resistance is thereby aff cted only very slightly (being reduced to about $99 \Omega$ ) and the effect of small tolerances in the $10,000 \Omega$ is quite negligible. On the other hand, the tolerance in the roo $\Omega$ is practically the same as that of the combination. In the parallel connection, then, the two tolerances are not independent of one another; the effect of varying one resistance depends on the value of the other.

Having neglected $p_{2} / 100$ in comparison with 1 , we should not expect to find this formula very accurate if applied to our $500-\Omega \pm 60 \%$ resistor. In parallel with the $200-\Omega \pm 1 \%$ resistor, the result is (by exact calculation) $142.8 \Omega+12.9 \%$ or $-30.2 \%$. (Note how much smaller the positive tolerance is than when they are in series, because the greater part of the current flows through the $1 \%$ resistor. This is much less so with the negative tolerance.) According to the approximate formula. $p$ is $\pm 17.86 \%$.

I have made rather a lot of this difference between the two cases because it forms an interesting analogy with intermodulation in amplifiers. If an amplifier is perfectly linear, then the effect of amplifying two signals at once is equal to the sum of the effects of separate amplification. But if, as is always more or less so, there is non-linearity, the signals intermodulate. And while this may be negligible when the amplitudes (relative to the steady valve current) are small, it will probably not be when they are large. The tolerances or errors correspond to the signal amplitudes.

The differential calculus method is a considerable help with the parallel circuit. Normally, for the benefit of readers who look on it as something much too intellectual, I keep the calculus out of these pages; but this time, for the benefit of those who have learnt the elements of it and are looking around for something not too hard to practise on, I will just stop to point out the way. It is based on the assumption that over the range of variation represented by the tolerance the "curve " is linear. In other words, exactly the same assumption as we make in most valve calculations. Using the same symbols as before, one can write it

$$
r=r_{1} \frac{\partial \mathrm{R}}{\partial R_{1}}+r_{2} \frac{\partial \mathrm{R}}{\partial \mathrm{R}_{2}}+\ldots .
$$

which means that to find the contribution of $r_{1}$ to the overall error one multiplies it by the rate at which $R$ changes when $R_{1}$ is varied; and similarly for any number of resistors connected in any way whatsoever. One only needs to know how R is related to the component resistances and how to differentiate with respect to each of these.

With two resistances in parallel the result of the

$$
\begin{aligned}
& \text { differentiation is } \\
& \quad r=r_{1}\left(\frac{\mathrm{R}_{2}}{\mathrm{R}_{1}+\mathrm{R}_{2}}\right)^{2}+r_{2}\left(\frac{\mathrm{R}_{1}}{\mathrm{R}_{1}+\mathrm{R}_{2}}\right)^{2}
\end{aligned}
$$

and when the substitution has been made for $p$, etc., the result is the same approximate formula as before.

The more complicated the problem, the greater the saving by using the calculus.

By one method or another we now have formulae for dealing with both series and parallel arrangements, and with these can solve any resistance network that can be reduced to such arrangements. And remembering that capacitances in series are calculated as for resistances in parallel, and vice versa, we can deal with them, too; which is rather useful when designing oscillator circuits for superhets.

In practice, one may have to take into account temperature tolerances as well as manufacturing tolerances; but the methods are the same.

## Meter Tolerances

When one comes to meters there is more need to consider the causes of the errors. Meter tolerances are generally stated as a percentage of full-scale reading; in other words, as a constant amount regardless of reading. For instance, if a $0-500$ voltmeter is said to have an accuracy of $\pm 1 \%$, it means that the readings are (or should be) reliable within 5 V . If one tries to read 10 V on such a meter, $p$ is $\pm 50 \%$, which is not very good. Moral : use a lower range. The only sort of error I can think of that would strictly follow this law is a faulty adjustment of the zero-setting device, and that doesn't count! Most of the inherent errors of the instrument -temperature, lack of balance, non-linearity of field and spring, etc.-have some sort of proportionality to the deflection; but it is not necessarily simple proportion, so to be on the safe side the maker quotes a full-scale error everywhere.

When using multipliers and shunts the plot thickens. Presumably any errors arising from uncertainty in the resistance of the meter are not included in the quoted inaccuracy and have to be added. So the question " A meter $\pm \mathrm{I} \%$ f.s.r. is used with a shunt $\pm 2 \%$; what is the accuracy of the reading?" is less simple than it looks. Assuming that the $\pm I \%$ refers to the meter used unshunted, we would have to know the maximum error in the specified or measured value of the meter resistance

Fig. 3. Symbols used in working out the shunting error due to tolerances in the shunt and meter resistances.


Fig. 4. Showing how a tolerance in a dimension is approximately doubled when the quantity is squared.

This would enable us to calculate the corresponding figure for the meter current in relation to the total current to be measured. We assume that this total current is unaffected by inserting the meter ; otherwise that error would have to be calculated, too. The I \% meter error is additional to the shunting error.
The symbols in Fig. 3 are, I hope, self-explanatory. A shunting-error formula can be obtained in much the same way as the parallel-resistance formula, but the starting point is different :

$$
\mathrm{I}_{m}=\mathrm{I} \frac{\mathrm{R}_{s}}{\mathrm{R}_{s}+\mathrm{R}_{m}}
$$

The result, either by straightforward algebra, neglect ing terms divided by roo, or by short-cut calculus, is

$$
p \simeq\left(p_{s}-p_{m}\right) \frac{\mathbf{R}_{m}}{\mathbf{R}_{m}+\mathrm{R}_{s}}
$$

This shows that if both meter and shunt resistance are high or low by the same percentage, the shunting error is nil (which is what one would expect).
An important thing to know is that if the value of a component has to be squared to give the answer, the overall tolerance is doubled. Similarly, cubing multiplies the tolerance by 3 ; a square root by $\frac{1}{2}$; and so on. This is another of the rules that is a good approximation for small tolerances only. It is very obvious to anybody who has absorbed the first stages of the differential calculus; and can be demonstrated to others by the simple diagram Fig. 4. The area A of a square is of course equal to $L^{2}$. If $L$ is increased by a small amount, $l$, A increases in the same proportion in two places, plus a little corner piece. If the
corner is neglected, the area therefore varies twice as fast as L.
The same sort of game can be played with a cube. but beyond that it is difficult.

If there is still space I would like to refer to a formula quoted on page 358 of Scroggie's " Radio Laboratory Handbook" ( 5 th edn.) as an example of a type that should be avoided for its insidious errormultiplying nature. It refers to the method of measuring the total initial capacitance of a tuned circuit by noting the frequencies $f_{1}$ and $f_{2}$, to which it resonates when two known capacitances, $C_{1}$ and $\mathrm{C}_{2}$, are added. The formula is

$$
C_{0}=\frac{C_{1} f_{1}{ }^{2}-C_{2} f_{2}{ }^{2}}{f_{2}^{2}-f_{1}{ }^{2}}
$$

Unless you have some time on your hands I don't advise working out the maximum error formula for this; but having done so I would like to give a typical example, assuming these values:
$\mathrm{C}_{1}=200 \mathrm{pF} \pm 0.5 \% ; \mathrm{C}_{2}=400 \mathrm{pF} \pm 0.5 \% ;$
$f_{1}=\mathrm{r}, 000 \mathrm{kc} / \mathrm{s} \pm 0.1 \% ; f_{2}=733 \mathrm{kc} / \mathrm{s} \pm 0.1 \%$.
Anybody using these figures might imagine that the data were pretty accurate; nevertheless, he would not be justified in relying on the answer to better than $20 \%$. Yes; decidedly a method to be avoided. Answer to Introductory Example

Apart from the $1 \%$ by which the nominal value of the parallel combination fails to fit the exact requirement, the tolerance in the $10-\mathrm{k} \Omega$ resistor could be (calculated by substitution in the given formula) II. $2 \%$. So a $\pm 10 \%$ resistor would do.

# APILIEIDELECTIENICS 

Brit.I.R.E. Presidential Address

SELLLING " electronics to industry, the dissemination of technical information, acronautical radio and the possibilities of employing vertically radiated transmissions in an endeavour to accommodate more stations within the limits of the h.f. broadcasting bands are some of the matters dealt with by Paul Adorian in his recent Presidential address to the British Institution of Radio Engineers.

He suggested that in order to stimulate interest in, and secure the acceptance of, electronic methods in industry and medicine, papers should be presented, not only to radio engineers but to those engaged in fields of research to which the techniques are applicable

In regard to the problems associated with broadcasting, Mr. Adorian summarized the difficulties, so far as medium- and long-wave stations are concerned, by saying that, allowing for a $9-\mathrm{kc} / \mathrm{s}$ separation between channels, it would be necessary to use each available channel twenty-five times to cover the world's land mass with one programme.

On the question of h.f. broadcasting, he outlined a system employing aerials which radiate a large proportion of their power vertically, giving relatively short-range reflections from the Heaviside layer. The system, using the wavelength of 30 metres, has been used in Trinidad since 1947 and has given satisfactory reception throughout the island in daylight, but at
night interference is caused by stations operating in different parts of the world on the same wavelength but with aerials radiating considerable power at low angles. Another station using the system has recently been opened in Jamaica.

## RCA Receiving Tube Manual

ANEW and completely revised edition of this useful reference book has recently been announced by the RCA Tube Department. It contains general information on valves, ranging from elementary theory to descriptions of the latest applications of receiving valves, and has some new features reflecting developments in electronics. The section on valve and circuit theory has been expanded and includes formulæ and examples for the calculation of power output, load resistauce and distortion for several classes of amplification, as well as information on the design of cathode followers. New designs for receivers and amplifiers have also been added, and there is a complete section on resistance-coupled amplifiers.

In addition, the manual provides technical data on more than 460 RCA receiving valves and television picture tubes, including many discontinued types. For quick and easy reference there is a classification chart which groups together the types having similar characteristics and the same filament or heater voltages.

Single copies of this new edition, RC-16, can be ordered from RCA Photophone Ltd., 36, Woodstock Grove, London, W.i2, price 5 s 6 d , post free.

# Lonǵ Rang̣e Television 

Review of Propagation Conditions
Prevailing During Reception in South Africa

By T. W. BENNINGTON (B.B.C.) and R. MORRIS (Panorama Receiving Station, S.A.B.C.)

REGULAR reports on reception of the Alexandra Palace $4 \mathrm{I} .5-\mathrm{Mc} / \mathrm{s}$ sound channel have been sent since 13th March, 1949, to the B.B.C. from the Panorama Receiving Station of the South African Broadcasting Corporation, near Johannesburg. At this station from 5 to 7 observations are made daily throughout the year on this frequency, covering both the forenoon and afternoon transmitting periods. The results of the observations are reported in a 5 -figure code, in which the first figure designates the signal strength on a scale of $0-5$, while the remaining figures indicate respectively the fading, noise, interference and overall merit. In this review we are concerned with the propagation conditions and so the signal strength figure only has been used. During all observations when it is not possible to allot a figure of 1 , or greater, to the reception, either "Nil" reports are given, or else the clesignation " BNL" (Below Noise Level) is used, both of which classifications are taken to mean that the signal strength is, in fact, zero.

The reports have here been analysed for the period 13 th March 1949 to 3ist July 1950. In the upper section of Fig. I are shown the days on which reception was obtained. On each of these days the observation giving the strongest signal of the day was taken and the results are plotted as vertical lines on the scale $0-5$. During 1949 reception was far more frequent, and signals were, in general, stronger during the months of March, April, October and November than cluring the other months of the year. It should be noted, however, that had observations been made during the early part of the year, frequent reception might have been had during February. Reception conditions were the most favourable in the early spring and late autumn, though midwinter produced occasional reception. During the
summer conditions were particularly poor and on one day only during June, and on not a single day in July, was reception obtained, though just as many observations were made then as during other months. In 1950 signals were heard on several days during mid-winter, though it again tended to increase towards the vernal equinox. It is again significant that not a single case of reception was obtained between 24th May and 3ist July when our record finishes. It is to be noted that reception during the early spring of 1950 was much less frequent than during that of 1949, due, no doubt, to the decreased solar activity and the consequent fall in the ionization of the $F_{2}$ layer.

In the lower section of Fig. I there are given, in histogram form, the total number of cases of reception at strength I or greater, as a percentage of the total number of observations made during the month. Reception on $4 \mathrm{I} .5 \mathrm{Mc} / \mathrm{s}$ was most frequent during March 1949, when 65 per cent of the observations produced signals, and it was least frequent during June and July I949, while June and July 1950 were completely blank months. It will be noted that in the late winter/early spring period reception peaked during the equinoctial month of March, whereas at the late autumn/early winter period the equinoctial month of September did not produce a notable increase in reception, which reached its seasonal peak in the following month. It would be expected that reception would peak just after and just before the mid-winter period rather than at the equinoxes, and it is difficult to assign a reason for this assymmetry in regard to the peak reception months of 1949, particularly as sunspot activity during April and September was at similar levels

It seems reasonable to assume that the best conditions for reception tended to occur towards both

Fig. I. Graphical records of reception of Alexandra Palace $41.5 \mathrm{Mc} / \mathrm{s}$ sound signals in Panorama South Africa. Day-to-day signal strength is given in the top half, the lower showing the monthly reception as a percentage of the observations made.




Fig. 2. Predicted monthly m.u.f.s for the Daventry/Johannesburg path at IIoo and 1500 hours G.M.T. during I949 and 1950.
vernal and autumnal equinoxes. Again in March and April 1950 there occured an increase in reception, followed by a sharp decrease in June. The level of reception in the spring of 1950 was very much less than in the same months of the previous year, which may be accounted for by the fact that the sunspot activity (as indicated by the monthly sunspot numbers) has decreased by approximately one-third as between March/April 1949 and March/April 1950.

## Comparison with Ionospheric Data

It is not feasible to compare these results with any measured m.u.f. (maximum usable frequency) data, because no such charts are published, and there do not exist enough ineasured critical frequencies over this path on which to estimate the results with the predicted m.u.f. for the path, as obtained from the world ionospheric contour charts, which are compiled on the hasis of past measured data supplied by the world network of ionospheric measuring stations.

It is not to be expected that, because reception was obtained on a few days during any particular montb, the mean m.u.f. for that month would reach 41.5 $\mathrm{Mc} / \mathrm{s}$. It is known that there is a very considerable day-to-day variation about the monthly mean value, and so it might be possible to receive $4 \mathrm{r} .5 \mathrm{Mc} / \mathrm{s}$ on several days during the month when the mean m.u.f. for the path was much lower. Nevertheless, when reception was obtained on 50 per cent, or more, of the observations made, it would indicate that the mean monthly m.u.f. for the path must have been near $4 \mathrm{I} .5 \mathrm{Mc} / \mathrm{s}$.

In Fig. 2 are plotted, for 1100 and 1500 G.M.T., the predicted monthly m.11.f.s for the path for 1949 and 1950, as obtained by the use of world ionospheric contour charts. It will be seen that there is a large seasonal variation in the predicted m.u.f. for the path, such that low values are to be expected in the summer, highest values before and after the mid-
activity, so that the late winter/early in spring peat in m.u.f. in 1950 is lower than either peak of 1949, and this agrees with the decreased amount of reception in the late winter/early spring of 1950 as compared with the peak reception periods of 1949. Actual values of predicted m.u.f. would appear to be somewhat lower than those which must have prevailed, for during no month did the predicted mean exceed $36 \mathrm{Mc} / \mathrm{s}$, whereas during several months 41.5 $\mathrm{Mc} / \mathrm{s}$ was received frequently.

In conclusion we can summarize :-
(r) Reception of $4 \mathrm{I} .5 \mathrm{Mc} / \mathrm{s}$ over the London/ Panorama transmission path was obtained; (a) frequently during the late winter/early spring and late autumn/early winter periods, (b) occasionally during the mid-winter period, and (c) not at all during the summer period.
(2) The seasonal variation thus disclosed is in exact accordance with the expected variation of m.u.f. over the path, as indicated by the predicted m.u.f. values. There is strong evidence, however, that the predicted m.u.f.s were somewhat lower than the actual frequencies for the path, at least during the equinoxes.
(3) Reception during the late winter/early spring period of 1950 was much less frequent than during both periods of peak reception of 1949, due to the general fall in sunspot activity and the consequent decrease in the ionization of the $F_{2}$ layer.

## INTERFERENCE MEASURING SET

THE Electrical Research Association has recently designed for the Admiralty a portable interference measuring set in conformity with the requirements of BSi 597 (suppression of marine installations) and BS727 (general measurement of radio interferance). Sets to this specification, which are equally suitable for marine or land use, are being made by E.M.I., though the Admiralty states other civilian organizations may use the design without change. Further information may be had from the E.R.A., Perivale, Middlesex.

# Phase-Shiit 0scillators 

New Circuit Giving Constant Amplitude

By W. G. RAISTRICK (Pye, Ltd., Cambridge)

THE circuits to be described are all based upon the phase-shift oscillator shown in Fig. I. In order to produce self-sustaining oscillation in this arrangement two conditions must be satisfied. First, the voltage introduced from the output of the amplifier must be in phase with that fed back to the input, and, secondly, the voltage gain in the amplifier must be rather greater than the loss in the resistancecapacitance network.

A further requirement, important when a variablefrequency oscillator is considered, is that the sum of the amplifier-gain and the network loss should be constant over the whole of the frequency range in order that the voltage output may remain constant and the amplifier valve work always on the same part of its characteristic. The success or otherwise of any RC oscillator with low harmonic distortion depends very largely upon how accurately one can maintain the system in a condition when it oscillates only very gently, and in practice most oscillators of this type embody some form of automatic amplitude control. The design of this control is greatly simplified if the amplitude variation in the first place is only slight.

If we make the assumption that the valve-amplifier gives a constant voltage-gain and a constant phaseshift of $180^{\circ}$ over the whole range of frequencies to be considered, then we are left with the problem of designing a phase-shifting network which gives a phase change of $180^{\circ}$ and a constant loss over the frequency range.

In the circuit of Fig. I it is not feasible to obtain a phase shift of more than $60^{\circ}$ in each resistancecapacitance section so that the minimum requirement for oscillation is three sections. If the network loss is to remain constant then either all the capacitors or all the resistors must be varied simultaneously and it is this latter feature which has led to the comparative neglect of this type of oscillator for generating variable-frequencies.

Three-gang potentiometers are not easy to obtain, and if the normal type of tuning capacitor of 500 600 pF , is pressed into service then at least two will be required for each section, if frequencies of the order of $20 \mathrm{c} / \mathrm{s}$ are contemplated, otherwise the associated fixed resistors become of so high a value that the valve grid is extremely susceptible to hum pick-up. A further consideration is that the loss in the network decreases with the number of sections. It is $\mathrm{I} / 29$ for a three-section network and $\mathrm{I} / \mathrm{I} 8$ for a four-section one, necessitating valve voltage gains of rather more than 29 and 18 respectively. This is one reason why some of the oscillators described in the past have had as their tuning elements two four-gang capacitors coupled together.

By employing the more elaborate phase-shifting circuit of Fig. 2 it is possible to obtain a phase-shift
per stage of approximately twice that obtained with the simpler circuit. As is well-known, the virtues of this particular phase-shifting circuit are that the phase may be changed from $20^{\circ}$ to $160^{\circ}$ without any appreciable change of amplitude at the output terminals, so long as no attempt is made to draw power from the circuit, and it thus becomes possible to design an oscillator with only two stages of phaseshifting. Also, only one variable element is required although, as will be pointed out later, it is not always advisable to avail oneself of this characteristic.

## Phase-splitting Valve

In all the oscillator circuits which have so far been tried by the author the centre-tapped transformer of Fig. 2 has been replaced by a phase-splitting valve. Each phase-shifting section then becomes as in Fig. 3.

For a first analysis we may neglect the effects of the valve anode resistance, the input impedance of the succeeding stage and the cathode-bias resistor.


The equivalent circuit is then that of Fig. 4 (a). If now a sinusoidal voltage of amplitude $E$ is applied between terminals $a b$, then the current through the resistive branch becomes $\frac{E}{2 R_{0}}$, and that through the RC branch $\frac{E}{R+I / j \omega C}$. The resulting voltage distribution is shown in the vector diagram of Fig. 4 (b) ; note in this that the ancle $a d b$ will remain a right angle as $C$ or $R$ is varied assuming $C$ to be a capacitor of low power factor, so that the point $d$ will describe a semicircle about the centre $c$, and $c d$ will always be equal to both $a c$ and $c b$ since all three are radii of the same circle. The voltage $c b$, the cathode voltage, is in phase with the input voltage to the valve grid, and $c d$ is the output voltage, so that the phase angle $\theta$ between input and output voltages is the angle $b c d$. By the process of completing the parallelogram, Fig. 4 (c), and bisecting angle $\theta$ by the perpendicular cf, it can easily be seen by inspection that the tangent of haif this angle is: $\tan \frac{\theta}{2}=\omega \mathrm{CR}$. Now $\tan 2 x=\frac{2 \tan x}{1-\tan ^{2} x}$ from which $\tan \theta=\frac{2 \omega \mathrm{CR}}{\omega^{2} \mathrm{C}^{2} \mathrm{R}^{2}-1}$
(1)

Fig. 5 depicts an oscillator constructed by associating two such phase-shifting stages with a valve-amplifier $V_{3}$ If we denote the phase shift obtained in the second stage by $\phi=\frac{2 \omega \mathrm{C}_{1} \mathrm{R}_{1}}{\omega^{2} \mathrm{C}_{1}{ }^{2} \mathrm{R}_{1}{ }^{2}-\mathrm{I}}$ then oscillation will be obtained when $\theta+\phi=180^{\circ}$.

Now $\tan (\theta+\phi)=\frac{\tan \theta+\tan \phi}{1-\tan \theta \tan \phi}$ and substituting from (I)

$$
\tan 180^{\circ}=\frac{0=\frac{2 \omega C R}{\omega^{2} C^{2} R^{2}-1}+\frac{2 \omega C_{1} R_{1}}{\omega^{2} C_{1}{ }^{2} R_{1}{ }^{2}-1}}{1-\left(\frac{2 \omega C R}{\omega^{2} C^{2} R^{2}-1} \cdot \frac{2 \omega C_{1} R_{1}}{\omega^{2} C_{1}{ }^{2} R_{1}{ }^{2}-1}\right)}
$$

$$
\text { From which } \omega^{2}=\frac{\mathrm{I}}{\mathrm{RCR}_{1} \mathrm{C}_{1}}
$$

$$
\begin{equation*}
\text { And } f=\frac{T}{2 \pi \sqrt{\mathrm{RCR}_{1} \mathrm{C}_{1}}} \tag{2}
\end{equation*}
$$

It now can be seen wherein lies the disadvantage in having only one tuning element, for assuming we decirle to allow the resistance R to fulfill this function then $C=C_{1}$ and the expression for oscillation frequency becomes $f=\frac{I}{\sqrt{\bar{R}}, ~ \sqrt{\bar{R}_{1}} \mathrm{C}}$; in other words the frequency becomes proportional to $\frac{I}{\sqrt{R}}$ and a linear control will give an extremely cramped scale if a coverage of the order of 10: I for each range is considered. In order to obtain an approximately


Fig. 4. Basic phase-shift circuit alone (a) and vector diagram of voltages (b). The parallelogram is completed in (c).

Fig. 5. Circuit of complete oscillator with two phase-shift and one amplifier stages.

logarithmic scale then a logarithmic potentiometer will have to be used and unfortunately logarithmic wire-wound potentiometers are not obtainable in large ohmic values.

However, when extreme accuracy and permanence of calibration are not important, quite useful instruments can be assembled around the ordinary carbontrack volume control with a $\log$ or semi-log law. A practical design along these lines would proceed somewhat as follows. The minimum value of $R$ should not be reduced below about $5 \mathrm{k} \Omega$, and from (2) it can be seen that the maximum value is equal to $\left(f_{1} / f_{2}\right)^{2} \mathrm{R}_{\min }$ where $f_{1} / f_{2}$ is the ratio between maximum and minimum frequencies on any one range. For decimal scales this ratio is, of course, 10 , and $\mathrm{R}_{\text {max }}$ becomes $0.5 \mathrm{M} \Omega$. R could conveniently be a fixed resistor of $5 \mathrm{k} \Omega$ in series with a $0.5-\mathrm{M} \Omega$ potentiometer. $R_{1}$ should be a fixed resistor equal in value to $\sqrt{\frac{R_{\max }}{R_{\min }}} \cdot R_{\min }=5 \times$ ro $\mathrm{k} \Omega$, and $C$ would equal $C_{1}$, and be simultancously switched for each range, increasing in capacitance in multiples of 10 . It will be found that sufficiently good multiplication will be obtained over the whole frequency range to enable a single calibration to be used, and the range-changing switch used merely as a multiplier, labelled $\times 1, \times 10, \times 100$, etc., so long as sufficient care is taken over decoupling arrangements at the lower
frequencics. The coupling capaciturs preceding the phase-splitting stages need not be embarrassingly large even for quite low frequencies as the input impedance of the particular type of phase-splitter used is high. A value of o.r $\mu \mathrm{F}$ is adequate for frequencies of the order of $20 \mathrm{c} / \mathrm{s}$.

The anode and cathode load resistors $\mathrm{R}_{0}$ should be of fairly low value-something between 2 and $5 \mathrm{k} \Omega$ will be found suitable. The "gain" of each phaseshifting stage is that which is normal for this type of phase-splitter and should not be less than o. 8 or 0.64 for the two together; thus the gain of the amplifier $V_{3}$ need not exceed 1.6 and its design becomes very simple

No provision for automatic amplitude control is made in the circuit of Fig. 5. The feedback is controlled by the $5-\mathrm{k} \Omega$ potentiometer and, once set, it will be found that reasonably constant output will be obtained over the frequency range without further adjustment. When the output is required to be as free from harmonic distortion as possible, then it must be set at the lowest possible setting consistent with reliable oscillation, and under this condition the total harmonic content can easily better i per cent.

Readers who prefer to embody some kind of amplitude limiting in their designs can use any of the usual arrangements. In this connection it may be mentioned that both valve-operated control circuits and tem-perature-controlled devices, such as Thermistors,
have been used by the writer with good results Enough has been said to indicate that the use of the valve phase-shifter* principle makes avaibable a RC oscillator allowing fair flexibility of design. A single- or a double-element control may be used as desired, and this control may be either resistive of capacitive ; for instance the circuit of Fig. 5 could have employed a single capacitor as the tuning control with a range of 3.16 to 1 , covering the audio-frequency band of $20-20,000 \mathrm{c} / \mathrm{s}$ in 6 switch-positions, Alternatively a twin-gang potentiometer can be made to give a range in excess of $30: 1$ in one sweep, and in all cases the output remains constant.

These advantages are secured at the expense of two extra valves, but this is not so great a complication when it is remembered that both phase-shifters can be sections of one of the popular small twintriodes, and $V_{3}$ can be another scction of a further valve of the same type. The complete oscillator plus output stage need therefore have no more than two " bottles." With the circuit of Fig. i the valve must nearly always be a multigrid type in order to obtain the necessary gain. In addition a cathodefollower is commonly added, so that the RC network may be fed from a low-impedance source, and two glass envelopes are still required, for the oscillator alone.
*British Patent No. 3516/1949

## B.B.C. REPORT FOR 1949/50

ACOMPREHENSIVE account of the work and progress in the operation of both the sound and vision services of the B.B.C. during the year ended last March is given in the Annual Report of the Corporation, which is published by H.M. Stationery Office.

After dealing with the engineering developments of the year, the section devoted to the technical aspects of the Corporation's activities, concludes with some notes on the television research undertaken. In order to obtain data concerning the probable mutual interference which would be experienced with synchronized television stations working on the same wavelength (which will, of course, be necessary as there are only five channels available and ten stations are planned) six recording posts were in almost continuous operation during the year and an unbroken record was kept of reception from experimental transmitters over long distances. Recordings were also made of the Alexandra Palace transmissions in Scotland and of the Sutton Coldfield transmissions in the London area.

As a long-term project to appraise the relative merits of different systems, it is stated in the Renort that preliminary experiments are being made on systems using higher standards of definition and on colour television. To facilitate work of this kind, a new flexible television transmitter designed to operate at will on standards of definition from 400 to 900 lines, has been constructed for use in the laboratories.

On the financial side, the Report records that although the Treasury retained 15 per cent of the licence revenue - $£ 1,753,926$-and the Post Office received $£ 8 \mathrm{I} 4$, III for expenses of collection and interference investigation, the Corporation's net licence income was $£ 9,938,917-$ an increase of $£ 494,445$ on the previous year. Of this amount $£ 272,747$ was derived from the additional $£ I$ charged for the combined sound and television licence. The revenue from the sale of publications was $£ 1,039,464$. It is noteworthy that 25.3 per cent of the year's expen.
diture on the Home and Television Services is classified as "engineering."

One of the best reproductions of an end-of-tube picture -showing Their Majesties at the Royal Opera House, Covent Garden-is among the illustrations in the Report, which costs 35 .

TELEVISION O.B. LINKS.-Six centimetre-wave transmitters, similar to that illustrated, are to be supplied to the B.B.C. by Marconi's to provide radio links for television outside broadcasts.


# Communications on $460 \mathrm{Mc} / \mathrm{s}$ 

Suitability of the Decimetre Waves for Mobile Services

By E. G. HAMER, B.Sc. (Eng.) (Hons.), A.M.I.E.E., Ass.Brit.I.R.E. (G.E.C. Research Laboratories, Wembley)

AFEW years ago it was thought that the frequency spectrum between 70 and $100 \mathrm{Mc} / \mathrm{s}$ would be wide enough to accommodate most of the necessary radio services, particularly shortrange fixed links and mobile communication. The demand for frequencies in this band for use in essential services such as police work has, however, been so heavy that additional allocations are not now readily available. Higher frequencies were therefore sought for some less essential services, including taxis and the press, and, at the 1947 Atlantic City conference the International Telegraph Union allocated the band between 156 and $184 \mathrm{Mc} / \mathrm{s}$ for this type of user. Here again, since only part of this band is available for commercial fixed link and mobile services, it appears that the congestion in this new band will shortly be the same as it is now between 70 and $100 \mathrm{Mc} / \mathrm{s}$. All the available frequencies are rapidly being allocated and the next logical step will obviously be to move yet higher up the frequency spectrum to the range between 460 and $470 \mathrm{Mc} / \mathrm{s}$, which was allocated at Atlantic City to fixed and mobile stations, and is used in America for "citizens' raclio."

Previous work has shown that frequencies in the 700-100-Mc/s band give slightly better results than those between 156 and $184 \mathrm{Mc} / \mathrm{s}$ when used for mobile types of service. For frequencies of the order of $460 \mathrm{Mc} / \mathrm{s}$, however, the performance has not been so accurately known. The various conflicting reports made in the past had led to the assumption that the usefulness of such frequencies for mobile communication largely depended on the location of the fixed station aerial.

## Mobile Station

In order to obtain more precise information on this point and, further, to investigate the behaviour of the received signals under non-" line-of-sight" conditions, an extensive programme of tests was undertaken in the London area. During these tests, the opportunity was taken of measuring the field strengths received at the mobile station and comparing these measured values with those predicted from theoretical considerations.

The tests were made using various fixed station sites, roughly corresponding to sites in open country, together with a site in the centre of a built-up area. The receiver was installed in a car, which was used as the mobile station and, in some of the tests, the transmitted power and the receiver sensitivity were adjusted to a value likely to be obtained with "walkie-talkie" type of equipment.
The transmitter consisted of a low power r.f. unit giving a frequency modulated signal at a frequency of about $\mathrm{II}_{5} \mathrm{Mc} / \mathrm{s}$. This unit was used to drive a series of doubler and earthed-grid
amplifier stages. A power output of 30 watts was obtained from the final amplifier and, with the length of the feeder in use, gave a radiated power of 6 watts from a simple vertical dipole aerial fitted with a quarter wave co-axial stub standing wave suppressor. The immense loss of power in the feeder with this arrangement was clue to the unavoidably large separation between the transmitter and its aerial. No attempt was made to increase the allround aerial gain, although at these frequencies an aerial array is very attractive owing to its small physical size.

A normal frequency modulated receiver was used in the car and operated on a frequency of $40 \mathrm{Mc} / \mathrm{s}$. This receiver was preceded by an extra frequency changing unit. Figure 1 gives a block schematic diagram of this arrangement, which consists of a capacity-loaded line connected to a silicon crystal valve used as a frequency changer. The local oscillator frequency is provided by a quartz crystal oscillator, followed by a chain of multiplier valves. An additional stage of amplification at the first i.f. of $40 \mathrm{Mc} / \mathrm{s}$ is included between the mixer head and the normal receiver. Such an arrangement gave a minimum receiver sensitivity for an intelligible output of $4 \mu \mathrm{~V}$ input across 70 ohms . A complete receiver designed in accordance with the latest practice might be expected to have a sensitivity of the order of $\mathrm{T}-2 \mu \mathrm{~V}$ across 70 ohms. The aerial fitted. to the car consisted of a quarter wave whip aerial mounted above a metal sheet acting as an earth plane. Again, no attempt was made to use an aerial array to give an increased gain.

Communication from the car to the fixed station was provided by normal v.h.f. equipment with a transmitted power of 10 watts and a receiver sensitivity of $\mathrm{I} \mu \mathrm{V}$ across 70 ohms. This circuit was operated at a frequency of $100 \mathrm{Mc} / \mathrm{s}$, so that an indirect comparison with the performance of the $460-\mathrm{Mc} / \mathrm{s}$ equipment was obtained by noting the areas in which signals to and from the car could not be received.

The characteristics of the equipment used can be


Fig. I. Schematic layout of the experimental $460-\mathrm{Mc} / \mathrm{s}$ receiver.
summarized as follows: (a) transmitter; radiated power 6 watts; peak deviation $\pm 45 \mathrm{kc} / \mathrm{s}$; operating frequency $466 \mathrm{Mc} / \mathrm{s}$; (b) receiver; sensitivity (for usable signal) $4 \mu \mathrm{~V}$ across 70 ohms; bandwidth $\pm 160 \mathrm{kc} / \mathrm{s}$ at 3 db points.
For the first series of trials, the fixed station aerial was locatec on top of a water tower near Ealing, izo feet above local ground level and some 200 feet above the surrounding terrain. The mobile set in the car was taken to a number of places, including non-"line-of-sight" locations behind near and distant hills. Field strength measurements were also made at selected sites, for subsequent comparison with the calculated values. The non-" line-of-sight" positions can be described as: (a) behind a long range of hills seven miles away; (b) behind a local hill two miles away; (c) behind a large hill four miles away, also partly screened by another hill between it and the fixed station.

The approximate service area for continuous reception had a radius of about eight miles, comparing very favourably with the roo-Mc/s signals. On good receiving sites, of course, much greater ranges were obtainable,

The second series of trials was made with the fixed station in the centre of London, the aerial being mounted 30 feet above roof level, the aerial height above ground level being 130 feet. Mobile tests were again conducted in and about the City of London, and field strength measurements were taken behind Hampstead Hill. Reliable communication in this instance was obtained up to a range of seven miles, and within a radius of two miles the effects due to fading of the signal were negligible.

## Some Comparisons

The results obtained in the two series of tests described above gave rise to the following conclusions: communication results obtained during mobile working on a frequency of $460 \mathrm{Mc} / \mathrm{s}$ are, in general, inferior to those experienced using a frequency of $100 \mathrm{Mc} / \mathrm{s}$. In isolated instances in built-up areas the $460-\mathrm{Mc} / \mathrm{s}$ frequency may give a better service owing to standing wave effect. As is usual with y.h.f. and e.h.f. installations, there is a variation in field strength depencling on multi-path propagation and, as a car moves through the resultant field, a fluctuation in field strength is encountered causing the strength of the received signal to vary. The rate of flutter is greater at $460 \mathrm{Mc} / \mathrm{s}$ than at $100 \mathrm{Mc} / \mathrm{s}$ since the wavelength involved is shorter. When frequency modulated transmissions are used, and provided the receiver has good limiters, the effect on intelligibility is of the same order as when using roo-Mc/s equipment. If, however, the limiting action of the receiver is poor, or the a.g.c. time-constant incorrect, this rapid flutter fading may give rise to an effect which is similar to a superimposed low frequency modulating the speech intelligence. The effect is to mar the speech quality badly and to reduce the intelligibility.

During both trials the equipment operating on $460 \mathrm{Mc} / \mathrm{s}$ did not give as great a range as that used on $100 \mathrm{Mc} / \mathrm{s}$. The difference between the two values was, however, usually small, being of the order of io to 20 per cent. It was also interesting to observe that, during communication to a heavily wooded country area, the trees appeared to cause more

Fig. 2. Ground contour of typical site used for field strength measurements.

attenuation of the $460-\mathrm{Mc} / \mathrm{s}$ signal than of the 100-Mc/s signal. This effect was more pronounced when the trees were wet, although the presence ot wet buildings in a built-up area appeared, if anything, to improve the $460-\mathrm{Mc} / \mathrm{s}$ communications.

The effect of operating the mobile station under trolley-bus wires and bridges was negligible, unless the bridge was of metal and, at the same time, very low and wide. Even in a town in very heavy traffic, the amount of ignition interference experienced was very small, despite the use of an intermediate trequency of $40 \mathrm{Mc} / \mathrm{s}$ and a relatively poorly screened experimental receiver. Good signals were obtained in the centre of the City of London about 2 miles from the fixed station, even in very narrow "canyon '"-like streets, and again the effects of passing traffic caused very little alteration in the signal. In particular, it was found that, in these narrow streets, there appeared to be no significant difference in signal level between adjacent streets which were radial and circumferential to the fixed station.

A further series of tests was carried out to simulate the conditions likely to be obtained using "walkietalkie" type of equipment, and in some instances the receiver used was of the simple super-regenerative type. The transmitted power was 0.25 to 0.5 watt, and the minimum receiver sensitivity for a usable signal was $30 \mu \mathrm{~V}$ across 70 ohms. From a high site the range to the sensitive receiver $\left(4^{\mu \mathrm{V}}\right)$ was $2 \frac{1}{2}$ miles in open country and 1 to $\mathrm{I} \frac{1}{2}$ miles in a built-up area With the super-regenerative receiver the range was reduced to 400 yards in a built-up area. In another instance the transmitter aerial was placed about 3 feet above ground level, adjacent to and outside the wall of a brick and steel building, halfway between the two ends of the wall. When using the superregenerative receiver inside the building the range obtained was between 60 and 100 yards, depending on the type of obstructions inside the building and upon the floor level on which the receiver was used. Outside the building (which was approximately 100 yards $s$ quare) signals were received along the side where the transmitting aerial was located and along the adjacent sicles, but they were not received on the opposite side of the building.

As a general rule the simple "walkie-talkie" equipment gives service, when used in the open, over a range of $200-400$ yards, depending on the location of the aerials. When used inside a building the range is reduced to 60-roo yards. It should, however, be remembered that at present it may not always be possible to achieve such performances using filament type battery valves.
An approximate analysis was made of the anticipated field strength in the areas where the field strength had actually been measured This analy-
sis was based on the nomograms published by Bullington.* $\partial$ For the non-" line-of-sight" position, the field strengths were calculated at the crest of the intervening hill to determine whether simooth earth or free space conditions should be used. The appropriate shaciow loss and, if necessary, the grazing loss were taken into account

Figure 2 shows the ground contour of a typical site tested. Allowance is made for the grazing loss over the intermediate hill, and for the shadow loss due to the principal hill. The calculated transmission loss is 119 d ), which is in satisfactory agreement with the measured loss of 124 db . A series of some 40 readings were made and the results were found to be consistent for any transmitting station site.

The results obtained with the transmitter on an open site show very close agreement with the results calculated by the Bulington method, the two values usually being within 3 db of one another for a simple non-" line-of-sight" path. This may, to some extent, be fortuitous in view of some of the assumptions made, particularly in regard to the gain or loss of the aerials and feeders. Using identical equipment in a built-up area, it was found that there was an extra loss of some 10 db in the measured values. This loss may be accounted for by the indeterminate extra
grazing" loss due to the roofs of surrounding build ings. Such extra indeterminate losses can be regarded as being a function of the transmitter aerial height above roof level rather than of its actual height above ground level. Small increases in aerial height are therefore likely to give much greater signals, while the converse is also true.

It can therefore be stated that, where a high trans mitting aerial is situated in the open, the propagation loss may be evaluated to a fair degree of accur acy, even under non-" line-of-sight" conditions. In a built-up area, however, an extra fixed loss must be added to the calculated loss. This extra fixed loss may vary in any given direction depending upon the heights and types of building near to the fixed station aerial.

The general conclusion drawn from the entire series of tests was that frequencies in the $460-470-\mathrm{Mc} / \mathrm{s}$ band could be used for many mobile radio services. A satisfactory service would definitely be obtained, although the service area would be slightly less than when using frequencies in the $70-100-\mathrm{Mc} / \mathrm{s}$ band. The service area could be increased by the use of aerial arrays giving omni-directional gain, these aerials being physically small, and might then be comparable with the $100-\mathrm{Mc} / \mathrm{s}$ service area. This general statement may not, however, be applicable in extremely rugged terrain so that, as when lower frequencies are used, exploratory tests on the actual site are desirable under such conditions.

[^3]
## Presenting Picknp Characteristics

THE need for a standardized presentation of audiofrequency response curves, to facilitate comparison, has long been recognized, and, as a first step, the Gramophone Equipment Panel of the Radio and Electronic Component Manufacturers Association have recommended their members to adopt a scale ratio of 38 db per octave. Graph paper on this basis (actually 3 inches per octave of frequency and $r$ centimetre to 5 db of vertical scale) has been prepared and is obtainable
from H. K. Lewis \& Co., 136, Gower Street, London, W.C.I. Tbe chart includes a useful ruled panel for relevant data

## Nomenclature: Standard Terms

SEVERAL supplements to British Standard 204:1943 $T$ ("Glossary of Terms used in Telecommunication") have recently been issued by the British Standards Institution, 28 Victoria Street, London, S.W.I

Supt. No. 2: Glossary of Terms used in Radio Propa. sation.-Terms connected with radio propagation through the ionosphere and troposphere. Price $2 s$.
Supt. No. 3: Fundamental Radio Terms.-Definitions of the terms used for the varions applications of radio in conununication and location, with a chart showing the relation between the various methods. Price is.
Supt. No. 4: Glossary of Terms used in Radar.Price 2 s .

## CLUB NEWS

Basingstoke.-The secretary of the Basingstoke District Imateur Radio Society advises us that owing to lack of support it has been decided to discontinue the Club's activities. Sec.: L. S. Adans, "Rosien," I6, Bramblys Drive, Basingstoke, Hants.
Belfast.-Two ryo-watt transmitters, one 'phone and one c.w., are now in use at the headquarters of the City of Belfast Y.M.C.A. Radio Club, Wellington Place, under the call sign Gi6YM1. Morse classes are held on Wednesdays-the club night-and Thursdays. Sec.: S. H. Foster (GI3GML), ${ }^{1}$ I, Belmont l'ark, Bellast.
Birmingham.-At the meeting of the Slade Radio Society in the Parochial Hall, Broomfield Road, Erdington, on December 8 th at 7.45 , W. H. Yeates (G.P.O.) will speak on "Telephone Transmission Systems." Sec.: C. C . Smart, (10, 'Voolmore Road, Ercington, Birmingham, 23.
Bournemouth.- Meetings of the Bournemouth Radio and Television Society (G3FVL) are held on the first and third Thursdays of each month at 7.30 at the Cricketer's Arms, Windham Road. Sec.: F. G. Hamshere, 99, Elmes Road, Winton. Bournemouth, Hants.
Brighton.-Membership of the Brighton and District Radio Club is now about So and meetings are held every Tuesday evening at 7.30 at the "Eagle Inn," Gloucester Road. The club station, G3EVE, is operated on alternate Tuesday evenings on 80 metres (c.w. and 'phone). Sec.: L. Hohden, ry, Hartington Road, Brighton, Sussex.
Chester.-Morse classes are held for an hour prior to the weekly meetings of the Chester and District Amateur Radio Society (G3GIZ) which are held at 7.30 on Tuesdavs in the Tarran hut in the Y.M.C.A. grounds. Sec.: R. C. Windsor, 17. Hough Green, Chester.

Coventry.-Fortnightly meetings of the Coventry Amateur Radio Society are held on alternate Monday's at 7.30 at the B.T.H. Social Club, Holybad Road. On December 4 th demonstrations will be given of aids to reception. Sec K. G. Lines ( G 3 FOH ), 142, Shorncliffe Road, Coventry

Dorking.-Weekly meetings of the Dorking and District Radio Society ( $\mathrm{G}_{3} \mathrm{CZU}$ ) are held at the H.Q., 5, London Road on Tuesdays at 7.30 . A lecture on valve technique will be given on December 1gth. Sec.: J. Greenwell, $\mathrm{G}_{3} \mathrm{AEZ}, 7$, Sondes Place Drive, Dorking.

Gravesend.-Weekly meetings of the Gravesend Amateur Radio Society are held on Wednestlays at 7.30 at the club headquarters, 30 , Darnley Road, Gravesend. Sec.: R. E. Appleton, 23 , Laurel Avenue, Gravesend.
Malvern-Meetings of the Malvern and District Radio Society are held on the first Wednesclay of each month at $7 \cdot 45$ at the "Foley Arms Hotel." A lecture on "Microphones and T.oudspeakers" will be given at the December meeting.

Wakefield.-The Secretary of the Wakefield and District tinateur Radio Society winh give members of the club "An Lntroduction to Frequency Modulation" at the meeting at Service House, Providence Street, at 7.30 on December 13 hh. Sec.: W. Farrar (G3ESP), "Holmgroft," Durkar, Wakefeid.
Walivorth.-Meetings of the Walworth (Men's Institute) Kadio Club are being beld each Weinesday and Fridav from $z$ to 9 at The Avenue School, John Ruskin Street, London, S.E.5. Sec.: I. Gibbs, 22, Caspian Street, Camberwell, S.E.5.

# The Hall Efiect 

## Its Application to the Measurement of the Flux Density of Magnetic Fields

IN Clerk Maxwell's time it was thought that the distribution of current in a network of wires or a fixed solicl conductor was unaffected-apart from initial transient induction effects-by the application of a constant external magnetic field. Explanations of the failure of experiments to disclose any movement of the current itself, in spite of the very considerable forces experienced by the conductor, were usually sought in terms of the " incompressible fluid" theories then current. They did not entirely convince E. H. Hall, ${ }^{1}$ working in John Hopkins' University, U.S.A., who held that even if the incompressible fluid entirely pervaded the conducting material, a lateral pressure, if not a flow, should be detectable.

His early efforts to detect changes due to a magnetic field in the equipotential points at the sides of metal strips carrying longitudinal current (Fig. I) were inconclusive until, at the suggestion of Prof. Rowland, he tried extremely thin conductors (gold leaf on glass). These gave positive results and he was able to show that the lateral galvanometer current was, in fact, proportional to the product of the main current strength and the magnetic field. Later the effect was shown to be inversely proportional to the thickness of the material. Thus the lateral change can be expressed $\mathrm{V}=R \mathrm{IH} / t$. Where $R$ is the "Hall coefficient" of the material, $\mathrm{I}=$ longitudinal current in amperes, $\mathrm{H}=$ magnetic flux density in gauss and $t=$ thickness of conductor in centimetres.

## Theories of Conduction

According to modern theories of conduction in simple metals (e.g., copper and silver), in which the current is due to a drift of electrons from atom to atom under the influence of an external field, the effect is explained if we assume that the electrons travel in straight lines under the electric field, but describe curved paths under the influence of the perpendicular magnetic field. There is a corresponding distortion of the equipotential lines which gives rise to a potential difference at the side contacts. (Fig. I (b).)

In other metals and semi-conductors the polarity of the Hall voltage may be of opposite sign, indicating that the current-carrying elements have the equivalent of a positive charge. These elements are not necessarily particles, but may be "holes" in the structure of a crystal lattice which are temporarily unfilled by electrons. When an electron moves into a vacant site, a " hole" is left in an adjacent part of the structure and this vacancy can travel through the material until it reaches the negative point of connection, when it will be filled by an electron. At the other end electrons are extracted, thus keeping up the supply of " holes."

The Hall effect in germanium is very high and the ratio of Hall coefficient to specific resistance, upon which the deflection of any current-indicating meter

[^4]connected to the lateral terminals will depend, is about 200 times that of copper. G. L. Pearson ${ }^{2}$ has shown that it is suitable for the direct measurement of flux densities up to 20,000 gauss, and has described a simple instrument, involving only a $4 \frac{1}{2}-\mathrm{V}$ battery, a


Fig. I. Illustrating the Hall effect. Current distribution (a) without magnetic field (b) with field perpendicular to the paper. Transverse equipotential lines are shown dotted.

Fig. 2. Typical circuit for measuring field strength by means of the Hall effect in germanium.


[^5]microammeter and resistances, which can be used in conjunction with a thin slip of germanium crystal for measuring, for example, the flux in loudspeaker magnet gaps.

The circuit used by Pearson is given in Fig. 2. Ranges of 5, 10 and 20 kilogauss are provided and the fourth position on the range switch is used to connect the meter across a 15 -ohm series resistor to check the main crystal current against a calibration mark. A potentiometer (" set zero") is used to offset any residual lateral current arising from the difficulty of soldering the leads accurately to spots of equal potential.

## In Commercial Form

In this country a similar type of instrument has been produced commercially by the British Thomson Houston Company, Rugby, and is known as the Type $G$ gauss meter. A five-way rotary switch gives ranges of 5 , Io and 25 kilogauss and has "Off", and "Calibrate" positions; there are separate calibration and zero-setting controls. The germanium crystal is in the form of a probe and is protected by a nonmagnetic sheath having external dimensions of $0.14 \times 0.035 \times 1.25$ inches; longer probes are a vailable. The polarity of the transverse field corresponding to positive meter readings is marked on the probe.

B.T.H. gauss meter, Type G, and (below) close-up of the probe, which is protected by a non-magnetic sheath.


## " RADIO LABORATORY HANDBOOK" NEW EDITION

VALUABLE information on all aspects of test and measurement is given in the fifth, revised, edition of this popular book, recently issued from our publishers at 15s. The author, who is well-known as a consulting engineer and as a contributor to Wireless World, writes knowledgeably from a long and varied experience of radio. He does so in a way that will appeal to the amateur and to the professional radio man alike, for he is concerned not only with commercial laboratory apparatus but with keeping down expenditure by constructing and improvising one's own equipment. After dealing with the principles of correct measurement, he goes on to describe at length the various types of instruments and how they are used, then makes helpful suggestions on the most suitable apparatus for equipping a laboratory. Considerable space is devoted to particular methods of measurement on components, amplifiers and receivers, and special attention is given to v.h.f. work. There is also a chapter of useful reference information, together with an appendix on the construction of bridges.

This edition has been revised and new material has been added. In particular, the pages on oscillators have been re-written to include the latest developments in RC oscillators, additional information is given on thermistors, valve-voltmeters and cathode-ray oscilloscopes, and the reference section is up to date.

## MANUFACTURERS’ LITERATURE

Microphones and associated sound equipment described in a set of leaflets from Lustraphone, 1.td., 84, Belsize Lane, London, N.W. 3
Electric Motors of the more popular ratings in an abridged list from Higgs Motors, Ltd., Witton, Birmingham, 6 .
Television Aerials and accessories described in an illustrated pamphlet from Validus Aerials, 69, Hornsey Road, London, N. 7.
Coilpacks are catalogued and suitable receiver circuits given in a brochure from Osmor Radio Products, Ltd. Bridge Viev Works, Borough Hill, Crovdon, Surrey.
Midget Paper Capacitors, type Wg9, described in a "Hunt's News" leaflet issued by A. H. Hunt, Ltd., Garratt Lane, London. S.W.I8.
Television Components illustrated catalogue, a vailable to manufacturers from the Plessey Company, Ltd. (Components Division), Ilford, Essex.

Communications Receiver, Eddystone " 740 " generalpurpose model described in a leatlet from Stratton \& Co., Ltd., Eddystone Works, Alvechurch Road, Birminghan, 3 r.
"Noise and Vibration in Industry," a booklet dealing with noise-measuring instruments, from A. E. Cawkell, 7. Victory Arcade, The Broadway, Southall, Middlesex.

Sound-level Recorder, high speed, and pH Meter, Type 1900, described in leaflets from Dave Instruments, Ltd., 130, Uxbridge Road, Hanwell, Lòndon, W. 7 .

Schering Bridge specification in a bulletin from Muirhead \& Co., Ltd., Beckenham, Kent.

Intercommunication system; the Ediswan Mark II Loudspeakerphone described in a brochure from the Edison Swan Electric Co., Ltd., I55, Charing Cross Road, London, W.C. 2.

Receivers and Radiograms for 195I; short specifications in a catalogue from Pye, Ltd., Radio Works,
Cambridge. Cambridge.

Television Receiver, Model $\mathrm{TU}_{142}$; a descriptive leaflet from E. K. Cole, Ltd., Ekco Works, Southend-on-Sea, Essex.

Sound Heads for magnetic recorders; brief details in a leaflet from Bradmatic, Ltd., Station Road, Aston, Birmingham, 6.

Earth Analyser for detecting earthing faults on electrical apparatus; a leaflet from Runbaken Electrical Products, $7 \mathrm{I}-73 \mathrm{~A}$, Oxford Road, Manchester, 1 .

# Dark-Screen Television 

Use of Tinted Implosion Guards

IN a television picture, the "blacks" are actually the cotour of the unexcited portions of the phosphors on the face of the cathode-ray tube as they appear in the particular conditions of external lignting which exist at the time of viewing. Consequentıy, if the external lighting is fairly mgh, considerable loss of contrast is noted in the picture, because the light reflected from the face of the tube gives a greyish look to the phosphors.
The obvious way of preventing this loss of contrast is to eliminate external light altogether, but this is often impracticable or undesirable. When the picture must be viewed under conditions of high external lighting it is possible to maintain an adequate degree of contrast by using a suitably tinted transparent light-filter in front of the cathode-ray tube to cut down the light reflected from the face of the tube

Such a neutral-tinted filter in the form of acrylic sheet is now quite often used in television receivers and it fulfils a double function. In addition to improving the contrast of the picture when viewed in a room not completely blacked-out, its strength in the appropriate thickness gives adequate protection to the viewer should the tube collapse.

The increased contrast is due to the fact that external light in the room passes through the filter, strikes the face of the cathode-ray tube and is reflected back through the filter again, whereas the radiations of the tube phosphors pass through the filter once only. This considerably reduces interference by reflected light.

At first sight it would seem that to obtain a picture of comparable brightness when using a filter to that obtained on a set with the ordinary clear guard, the tube brightness would have to be turned up considerably to make up for the light absorbed in the filter. This is true to a certain extent, but in practice the actual colour of the filter is not a true "straight line" neutral and although the overall transmissions of the colours used are on the average about 50-55 per cent, their transmission curves show a considerable increase in transmision in the blue and red ranges. Transmission curves for the three " neutrals" which have been specially developed for this application are given in Fig. I, and show that the transmission rises in the blue to about 70 per cent and in the red to over 80 per cent. As the visible light from the excited phosphors of the average cathode-ray tube also peaks in the blue and red, the filter transmits about 70 per cent of the phosphor radiations as against only about 50 per cent of daylight or other external lighting.

A further point which must be remembered is that any type of clear guard of glass or plastic transmits at the best only about 90 per cent of the light from the cathoderay tube falling on it, so that the com-


Fig. I, Light transmission of "Perspex" acrylic sheet television implosion guards. The spectrum colours are indicated approximately against the wave-length scale.

This photograph shows identical sets under conditions of fairly high external illumination. The left-hand set has a clear guard and gives a picture of poor contrast; the right-hand set has a neuiral tinted "Perspex" filter and produces a picture of gcod contrast.

parison of transmission of a tinted to a clear filter is about 70:90 and not 70:100.

A further effect of the coloured filter is that it completely alters the "dead pan" appearance of the cathode-ray tube and mask.
The photograph-taken in the Pye development laboratories, Cambridge-shows two identical sets operating side by side under conditions of fairly high external illumination. The left-hand set, which has an ordinary clear guard, demonstrates a picture
of poor contrast. The right-hand set, which has a neutral tinted "Perspex" acrylic filter, manufactured by I.C.I. Plastics Division, England, shows good contrast. Behind the sets are exposure squares ranging in colour from white to black which demonstrate that the photograph shows the effect as nearly as possible as it is observed by the eye. These sets were, of course, individually adjusted to give as good a picture as possible under the existing circumpicture

# Radio in Germany 

Broadcast Receivers and Sound Reproduction at Dusseldorf

FORMING an opinion of the broadcast receivers on show at the first German post-war radio exhibition, held at Dusseldorf in August, was not made any easier by the scarcity of circuit diagrams, but in spite of this one gained the impression that modern German design techniques are much the same as ours, and no unusual principles are involved. In outward form the sets compare very favourably with models from other countries, and have neat and attractive cabinets without too much emphasis on originality of appearance. At the same time, there is a tendency in some quarters to camouflage receivers as objects of furniture. One superhet, for instance, takes the form of an easy chair, with controls fitted on the arms and a loudspeaker in the back facing outwards. The built-in frame-aerial gives somewhat directional reception, so the chair is provided with wheels to permit easy positioning. Another example is the " Straight Three Bedside Lamp," which has nothing in its outward appearance to reveal that it is a receiver, except that the station names are printed on the shade and a shadow line indicates the station to which the "lamp" is tuned.
Most of the medium-priced German receivers now have an e.h.f. band ( $86.5 \mathrm{Mc} / \mathrm{s}$ to 10 r $\mathrm{Mc} / \mathrm{s}$ approx.) incorporated in the design, and if not, arrangenents are made for adding a separate e.h.f. unit by plug-and-socket connection. Some makes also provide for i.m. reception, as there are several f.m. transmitters in use now, in Bavaria and N.W. Germany.

## Interest in Tuned Circuits

When judging the merits of a receiver, the German public is interested not only in the number of valves that go with it (usually five or six) but in the number of tuned circuits in the r.f. stages, and this habit of counting tuned circuits (an i.f. transformer counts as two) is reflected in all the sales literature. "Seven tuned circuits" is a sales point, so that a dealer wish ing to do business with only three has a difficult tine. However, the importance of the properties of the tuned circuits is also appreciated, and quite a few receivers at the exhibition had resonant circuits with adjustable characteristics-variable bandwidth, for instance. With some models, the bandwidth auto-
matically contracts on reception of a weak signalthereby giving selectivity and an adequate signal-tonoise ratio in exchange for a loss in quality-and then expands again when a strong station is tuned in. Other sets are fitted with a manual bandwidth control, and one manufaoturer has combined this with a tone control.

The average price of broadcast receivers in Germany is much the same as in this country, but the lowest prices are well below those charged here. For instance, a straight a.c./d.c. set for reception of local stations can be bought for 65 Deutsch-Marks (just over 5 , guineas), whilst the "Straight Three Bedside Lamp" mentioned above costs about the same. These low prices may be partly accounted for by the fact that the number of German manufacturers producing receiving sets has greatly increased since 1939 and competition is more intense.

## Unusual Recording System

Among the exhibits of the German Federal Post department, the most up-to-date equipment on show was a telephone amplifier using germanium crystal triodes, or transistors. Compared with the 8 watts of non-utilized power consumed by an equivalent valve amplifier (for heater supplies, standing anode current, etc.) this amplifier required only 0.16 watta saving in power or efficiency of 98 per cent.
In the acoustics section, a long-playing gramophone recording system was displayed, giving a smaller average spacing between grooves than is normally possible. With conventional systems of recording, the width of the grooves and the spacing of any two neighbouring grooves is kept large enough to accommodate any sound amplitude-that is, any lateral needle swing-that may be necessary in a recording, but in this new system the width and spacing are arranged to be dependent on the amplitudes which actually do occur in the individual grooves and neighbouring grooves. The result of this more economical method of "packing" is that more grooves can be got into the space available, and it is claimed that this gives an average increase in playing time of 100 per cent. Speech necordings would probably allow a greater increase in playing time than would musical performances.
M. L. T

# Manuiacturers’ Products 

## New Equipment and Accessories for Radio and Electronics

## Miniature Three-Range Coil Pack

A
THREE-RANGE coil pack measuring $2 \frac{1}{2} \mathrm{in} \times 2 \frac{3}{4} \mathrm{in} \times 1 \frac{5}{8} \mathrm{n}$ and intended for use in small superheterodynes without r.f. stage has been introduced by British Distributing Co., 66, High Street, London, N.8. When tuned by $0.0005-\mathrm{mfd}$ capacitors the three ranges available are: 16 to 50 m , 190 to 550 m and 1,000 to $2,000 \mathrm{~m}$ respectively.

The six coils have dust-iron cores. The pack is inductance-trimmed in the factory, and the only adjustments needed after its assembly in a receiver will be to the six trimmer capacitors accessibly mounted on top of the unit. With the single hole fixing and simple wiring, assembly takes only a few minutes. It is designed for an i.f. of $465 \mathrm{kc} / \mathrm{s}$. The Bridisco sub-miniature coil pack is priced at $30 s 6 \mathrm{~d}$

## Quality Amplifiers

ARANGE of quality amplifiers costing approximately $\mathcal{L I}^{1}$ per watt has been introduced by the Broadcast and Acoustic Equipment Company, Tombland, Norwich Resistance-capacity coupling is employed with a paraphase-coupled push-pull output stage

In the 12 -watt amplifer, 6 V 6 of 6 L 6 valves are used in the output stage and a 120 mv input is required for full output. Feedback is applied over three loops, and the response is flat within $2 \frac{1}{2} \mathrm{db}$ between $20 \mathrm{c} / \mathrm{s}$ and $15 \mathrm{kc} / \mathrm{s}$. Variable

bass and treble boost is included with maximum lifts of 20 db . The price is $\neq 12$ LOs, or $t^{8}$ los without power pack.

The 15 -watt model has 61.6 or 807 output valves and the overal sensitivity is higher, being suitable for use with a microphone or photo cell. The frequency range is $10 \mathrm{c} / \mathrm{s}$ to $20 \mathrm{kc} / \mathrm{s}$ within I .8 db . Bass and treble tone controls are included and the price is $£ 15$ Ios or $\notin 10$ ros with. out power pack

Breeze Plug and Socket for heavy-duty applications has two roo-amp pins and one 19 -amp pilot pin, in a brass housing that is waterproof when screwed together. All three socket inserts and the two ioo-amp pins can be removed for soldering. The makers are The Plessey Company, Ltd., of llford, Essex.

High-Voltage Adaptor, Model 34 I , has been introduced by Taylor Electrical Instruments Lid., 1 19-424 Montrose Avenue, Slough, Bucks, to extend the voltage range of their Model ifo A Electric Testmeter up to IokV

Television Receivers. Suital,le fot viewing in normal room lighting is the 600 A table model projection television receiver prodaced by Philips Electrical, Ltd., of Century House, Shaftesoury Avenue, V.C. 2 It has a fiat screen measuring $13 \frac{3 i n}{4}$ $x$ rolin with lenticular rulings on the black face to give a wide horizontal angle of view, and is fitted with safety devices to prevernt dam age to the projection tube should


Bridisco sub-miniature coil pack.
Ferranti Model Ti505 combined radio and television receiver, with self-contained radio aerials.

receiver $\mathrm{r}_{1205}$ two more 12 -in models, the Trqo5 table model, and the Ti505 combined radio and television receiver. There will be two versions of each model available, and Ferranti will incorporate converter mits, free of charge, in all Sutton Coldfield models which may have to be converted as a result of the forthcoming opening of the Holme Moss transmitter.

Two-speed Gram Motor. Coinciding with the recent offer to the public of long-playing records. Small Electric Motors, Lti., Eagle Works, Churchtields Road, Beckenham, Kent, have produced a twospeed gramophone motor operating at 78 and $33 \frac{1}{3}$ r.p.m. The complete motor assembly, which is of the rim drive type, is compact and suitable for mounting direct on to a metal or wooden baseboard. Although at the moment the standard unit is for use only on $50-\mathrm{c} / \mathrm{s}$ a.c. mains, a $60-c / s$ unit can be offered for the export market.

## Radio and Road Hogs

$\mathrm{R}^{\mathrm{E}}$ECENTLY I was telling you about a small radio transmitter installed in my car which, without infringing the law, enables me, a mile or so from home, to flood the house with light, warmth, music and the smell of sizzling kippers.

A reader, who signs himself "Aurophile," which, on the face of it, sounds like another name for Narcissus but probably means a car lover, has written to tell me that there is a far more necessary use for a short-range e.h.f. car transmitter than panclering to the soul-destroying sybaritism in which I indulge. He is apparently troubled by the number of inexperienced divers and experienced road hogs who hurtle along darts country roads at night with dazzling headlamps which they haven't, respectively, the savvy or the courtesy to dip when passing other road users; in fact, according to him, night driving has become a regular battle of headlights. I cannot confim or deny his words from ny now experience as I am at the present time far too engrossed in trying to navigate my car solely by
radio beam from an approaching car was picked up by a simple tixedtrequency receiver which would operate the necessary switch by a relay.

## Time Gentlemen, Please

Nationalization of coal is a burning topic, but, being politically controversial, is naturally barred, together with discussions of other nationalization matters, from the pages of Wireless W'orld. I do think, however, that the Editor may make an exception when I call attention to the shameless manner in which the Government's now ser* vants are deliberately sabotaging its most cherished mononolies. I refer to Time, which has had the offices of its control board at (ireenwich since vesting day. Wireless, and in particular the B.B.C., has made us all conscious of the boon and blessing of this particular piece of nationalization.

Yet one need only go into any nationalized railway refreshment room to see B.P.T. (British Pub Time) still rigidly observed and the Government's wishes flouted by its
of the self-contained type using a magnetic wire or tape from which the programme can be easily and quickly wiped out when desired. I have for a long time used a bomemade one of this tyne and I make a regular habit of recording certain 33.B.C. programmes just for the pleasure of wiping them out without listening to them.

However, as $I$ have said on a previous accasion, the real function of these instruments, in my opinion, is to record programmes which we cannot otherwise hear. Next year during the Festival of Britain when, as we have all been bidrlen, we are devoting all our energies to enlightening the overseas visitors on "this realm, this England," we shall have still less time to listen. We shall, in fact, have to bottle not only the B.B.C.'s day-time programmes but the evening ones as well for consumption in what we shall with literal truth be able, with apologies to Thomas Hood, to call "The Stille Night."

It is possible, of course, to bottle any programme at will by means of an ordinary receiver, a time switch and a recording unit. We can, in

its radar screen and accordingly I leave lighting and all such matters to Mrs. Free Grid, but possibly my correspondent's information explains the presence of the ex-A.A. searchlight that I notice has recently been mounted on the car roof.

Briefly, my correspondent's idea is that headlamps should all be made illegal and the latest type of overhead liglrting installed on all roads throughout the land, the capital cost being met partly by the money iaved by not manufacturing headlamps and partly by one of the Chancellor of the Exchequer's once-for-all levies. Running costs, represented chiefly by the huge current consumption, could be virtually wiped out by using radio since the lights on any particular furlong of road would not come on until a
own lackeys who fing thirsty electors out on to the cold hard platform five minutes before the legally appointed time. This five-minutesfast rule must irritate teetotallers also, for after gulping down Government tea and the very antly named rock cakes, they rush out on to the platforin only to find to their chagrin that they have plenty of time to get sedately into the third-class accommodation for which they have paid instead of llinging open the door of a moving first-class compartment and collapsing breathlessly into its cushioned comfort.

## Home-bottled Programmes

AM glad to see that more and more home recorders are coming on to the market, especially those
fact, select and bottle all our listening for a week ahead for consumption at our leisure. I am able, by this means, to keep a car full of guests cheerful and free from the vice of backseat driving when rolling home in the small hours long after the B.B.C. has closed down.

I have little doubt that by doing this I am not only infringing copyrights galore but also tiansoressing against the rights of innumerable recording companies; [ trust they will all send along the necessary writs to the Editor who will accept service on my behalf. I can unly think that it is this question of copyright and recording rights that prevents manufacturers putting forward an a $: l$-in-one receiver-cum-automatic recording unit. Can some of you legal luminaries enlighten me?

## LETTERS TO THE EDITOR

## The Editor does not necessarily endorse the

 opinions expressed by his correspondents.
## Resistor Colour-coding

Is$S$ it not time that the manufacturers of colour-coded resistors made up their minds as to the meanings of the colours? There is great confusion between Black for Nought and Brown for One-I have before nie two resistors with orange body and tip, and brown spot; on a bridge they are 33 ohms and 330 ohms respectively! Green Black Black is shown on the bridge to be 50 ohms-how would 5 ohms be marked, then? Among a batch of 50,000 -ohm resistors, the honours are about even betwcen Black and Brow'n for the tip colour. Surely, if Black means Nought, when it appears as body or tip it must mean Not Even a Nought (a simple explantation becomes a little Irish, but you must see what 1 mean).

Worse is to come-with certain high-stability resistors a fourth stripe appears for the tolerance and bears its correct colour (in place of the familiar gold and silver of the humdrum resistor). Some resistors have a black moulded body; here is one with brown tip and yellow spot; it measures 100,000 ohms, using the body colour for the initial ONE. But here is one with black body and two red stripes; it measures 2,200 ohms, showing that this time the black booly forms either the intermediate or final colour, according to taste; finally, a brown moulded booly with three stripes, Red Red Brown; this measures 220 ohms, showing that the body colour is now disregarded -you have my sympathy if you thought that this was a 1,200 -ohm I per cent resistor.

No wonder one manufacturer takes no chances and places round his colour-coded product a label bearing the value in claatly printed figures!

PETER D. DAW.
London, S.E.i.

## Stereophonic Broadcasting

$I^{\mathrm{N}}$N your issue of September, r950, I read an article about improved stereophong in which E. Aisherg started his article with the following words:
" The first broadcast of stereophony. the system in which sources of somind are restored to their relative positions in space, took place in France on June 19th, 1950.'

Although I understand the feeling of stating something new, in this
case there is a deus ex machina in Holland.

On June 15 th, 1946 , there was a stereophonic transmission with a specially constructed artificial head, equipped with two condenser microphones, that broadcasted a concert given by the Dutch RadioPhilharmonic Orchestra.

Also the manifold publications of our industry-Philips, Eindhoven, about stereophony (Dr. K. de Boer), were not mentioned at all in this publication.
J. J. GELUK,

Netherlands Broadcasting Union.
Hilversum, Holland.

T
HE article by E. Aisberg in your September number was of considerable interest to me, having had experience of an opposite problem during the earlier stages of the late war with acoustic locational apparatus for aircraft.

In his seventh paragraph he mentions the differential intensity of sound bejng responsible for the binaural effect, Although the intensity effect exists over short periods with human ears, it was always my experience that it was the phase differential aspect which predominated in any locational exercise or practice. To be equipped with tubes and receiving horns extending the aurab sound base from bin to 8 ft and to centre these upon a stationary sound source prorluced a remarkable effect in that the nodal point of no phase ditference travelled recognisably round the back of the cranium. This, of course, operated in one plane only for one "listener."
For perfection we will need a wall in which are positioned 5 loudspeakers (the middle one to take un) slack) and two listeners, one sitting normally to recognize what is going on in the horizontal plane, and one prone to follow the vertical movements of those who climb staircases or rise upon piano wire, or to hear, for example, the gravitational descent of "Les Larmes lu Diable."

> M. F. L. FALKNER.

Holmes Chapel, Cheshire.

YOUR contributor's account (September issue, page 327) of the recent French broadcast over two channels dues not, to my way of thinking, make a clear enough distinction between true binaural stereophony (in which two microphones mounted in an artificial head


Can you provide a public address system at a moment's notice? With a B65 it is simple-just place the equipment in a suitable position and switch on. Incorporated within an easily portable case are the amplifier complete with loudspeaker, rotary transformer, 6 -volt unspillable accumulator and microphone with cable. Power output is approximately 5 watts. The equipment is a most usetul outfit for political meetings, religious gatherings, auctioneers, etc., and numerous other applications where no electric supply mains are available.

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Irixadio Westo ondon.
are used at the source, and headphones are used at the receiving end) and other systems which simulate directional effects

Neither is it true to say that the appreciation of acoustic depih is essentially a product of binaural bearing. Distance between sound source and microohone in a studio is readily judged in a single-channel system by the ratio of direct to reverberant sound. The inclusion of a reverberation chamber in the system described suggests that the designers are well aware of these facts, and, together with two-channel control of volume, they have j)laced at the disposal of producers a versatile mediam of expression. I believe that Bernhart and Garrett describe their work as "directed" stereophony, which seems a happy phrase. HENRY MORGAN.

Findhead.

WITH reference to the article in the September Wiveless World, I note that there is no reference to the effect of phase displacement in the location of a source of sound.

I had understood that whereas amplitude formed the chief discriminating factor in the case of the high notes, phase discrimination was important in the lower register where amplitucle differences are less marked.
(This is physiologically possible since at low frequency the nerve impulses are synchronous with the sound waves. Note that this does not involve phase cliscrimination between waves of different frequency.)

If this is so Fig. I (a and b at least) will preserve the original phase relations intact, provided the channels are balanced, but Fig. 2 ought to introduce a certain amount of distortion unless the faders can be made to introduce the correct phase displacement, a problem of no small magnitude in an electrical circuit.
H. R. A. TOWNSEND.

South Cerney, Glos.
vision. The answer, of course, is an extension speaker, but this not only adds to the overcrowding, but is an additional expense. In the television set, however, there is a loudspeaker which is idle for a large part of the day. May it be sug. gested to television manufacturers that a simple switching apparatus fee incorporated in television sets (other than combined sound models, of course) so that the speaker may be used as an extension speaker from a wireless set. The speaker could be wired for use when the television set is installed. D. R. BRAY.

London, N. 22 .

## Raising E.H.T. Voltage

IFEEL the greatly improved brightness and focus of modern feceivers is due to using voltages of 6 to 10 kV whereas it was quite common in older types of receiver to use from 3 to 4 kV .

Tlae olsvious method of changing the e.h.t. transformer for one of higher voltage might be considered too costly in an old receiver. The only alternative is to use line flyback e.h.t. but it will usually be found that the line output trans. former is unsuitable for this purpose and with a simple half-wave system would only provile about 2 to 3 kV . But if this voltage could be added to the existing system the e.h.t. could then be 6 to 7 kV or more. Unfortunately, however, the straightlorward connection of a rectifier and condenser would not work as the cathode of the rectifier would he at a higher positive potential than the flyback pulse feeding the anode. The best way of adding the voltages is as shown in my diagram. This circuit might seem a bit puzzling until it is realized that the extra components $\mathrm{V}_{2} \mathrm{C}_{2} \mathrm{C}_{3} \mathrm{R}$ are really the top half components of the usual voltage doubler system sometimes used in flyback e.h.t. systems.

During the forward scan $C_{2}$ will be charged up through $R$ to the same potential as $C_{1}$ : When the

Hyback pulse comes along these two voltages will be added and will be applied to the anode of $V_{2}$, thereby charging $C_{3}$ and $C_{1}$ to the combined voltage. So therefore with the simple addition of two condensers, a resistor, and a rectifier, almost double the e.h.t. voltage could lee obtained. This would naturally mean that increased scanning power would be required and whether this is within the capability of the set is another matter, although $I$ have found this just possible in most cases.

It would be rather more difficult using the voltage doubler system with the existing mains transformer as one side is usually earthed and must remain so. In any case, much larger capacitors would be required whereas with the above method $C_{2}$ and $C_{3}$ need only be o.0oo3 $\mu \mathrm{F}$; R is 2 MI .

It would be better to use a metal rectifier for $V_{2}$ and no heater voltage would then be required.
L. J. HILLS.

Belvedere, Kent.

## American Insularity

$\mathrm{H}^{2}$AVING read with interest the correspondence on the above subject in your journal, I felt it would be interesting to analyse the footnotes and references in a typical Annerican publication. This I have done for the excellent collection of articles entitled ' Electronics Manual for Radio Engineers" (Zeluff and Markus; Pıh. McGrawHill). This book contains 872 pages o! text, covering 289 " all-timegreat articles" from Electronics 1940-48.

The total numbers of separate references are as follows: American, 466; British, 38; others (mainly German), 14; (A few references are of doubtful origin and may have been added to the wrong list). Of the 38 British references, 6 are to O S. Puckle's "Time Bases," 5 to Wireless World and 3 to Electronic Engineering, leaving a paltry 24 to

## Hint to Manufacturers

WHEN buying a television set its position in the home should be in a room different from that in which a wireless set is kept. This is obviously desirable when only one wireless set is available, as a member of a family may well wish to hear a particular sound programme during television transmitting hours. It is also desirable in these days of flats and small rooms which quickly tend to become overcrowded. At the same time sound broadcasting is sometimes required in the room used for tele-

represent the remainder of British brains.

A further analysis shows that of 77 pages of microwaves (did we do any radar work in this country?) there are 3 British references, and of 75 pages on television, only I! In an article on the "Phantastron," although it is admitted that the circuit originated at T.R.E., there are no British references and 3 American references.
It is important to realize that these figures are not taken from one isolated case, but represent the attitude of 270 different authors to the efforts of workers outside their own country.
Although I have not malle a similar count in any British publications, a quick check through 3 copies of Wireless World and 3 of Electronic Engineering shows that more than 50 per cent of references are to American literature
A. T. COLBECK.

Talgarth, Brecon.

## Names and Titles

$\mathrm{A}^{\mathrm{s}}$$S$ an "old timer" in the world of wireless I have been giving some thought to nomenclature and titles.
The first refers to v.li.f. radiotelephony between private velicles, tugs and the like, offices, etc., with, at present, the unwieldy designation of "Business Radio." It is extraordinary how difficult it is to arrive at a suitable name covering this application of the art. Of course there is $R / \Gamma$, but this encroaches upon service terminology and in any case is still not a word. RADIOPHONY appears to cover it, but it would be interesting to have opinions of your many readers.
My secon! search has been for a suitable title covering the 1951 Radio Exhibition, observing it cannot be held at Olympia. RADIOSHOW is what it is, but RADIOBRITAIN seems to cover the special occasion. Again, what do your readers think?
H. ANTHONY HANKEY.

London, W.f.

[^6]
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By "DIALLIST"

## Big Tube or Projection?

Interesting to notice that television receivers seem to be developing along quite different lines in this country and in the United States. Viewers in both countries want big images: here the tendency is to provide them by projection methods; on the other side of the Pond they are all in favour of the monster c.r.t. I don't know of any model amongst this year's British receivers with a tube bigger than 15 inches, nor are there many such; but in America there are quite a few with rg-inch iubes and for all I know some may use even larger ones. One reason, I suppose, for our different approaches to the problem of the big inage is that America, with an output that runs into quasiastronomical figures, can turn out c.r.t.s a good deal more cheaply than we can. We, on the other hand, seem to have advanced rather further in the technique of making reasonably long-lived, small, superbrilliant projection-type tubes.

## Direct Comparison

Not all comparisons are odious, despite the old saying. I recently had the opportunity of making one that can't have been possible to any vast number of people, and, far from being odious, it was surprisingly interesting and instructive. Briefly, I watched the simultaneous display of a whole television programme by projection and big-tube methods. The images were of the same sizeabout 16 by 12 inches-and the two screens were side by side. The big tube was provided with optional spot-wobble; that is, the "spotwobbulator" could be switched on or off at will. The first thing to strike one was that, even at short range, no lininess was apparent in the projected image, though it was very much in evidence in that on the large tube when the spot-wobble was not applied. The reason, presumably, is that in small high-voltage tubes one cannot obtain a focus that is relatively as sharp as that of a high quality tube of large diameter. Ir other words, the spot on the small tube has sufficient spread to cover up the effects of lininess. One might expect this to lead to a rather
poorly-defined image; but I cannot honestly say that it had this effect. My general impression was that thongh the projected image might be a little lacking in depth when directly compared with the other, it was distinctly more pleasing (and less tiring) to the eyes than an unwobbled image on the big c.r.t

## Passing Thought

It must be no mean job to design and manufacture very large cathoderay tubes. It wouldn't be so bad if the business end could have a pronounced convex curve, for this would have a much better chance of standing up to the strains resulting from a vacuum within and an outside atmospheric pressure of getting on for 15 lb to the square inch, but the end of the television c.r.t. must be as nearly flat as possible. Take a 20-inch tube to make calculations nice and easy, and you have a surface area for the screen of 314 square inches. That multiplied by 14.5 gives a total air pressure on the screen of 4.553 lb , or a little over 2 tons-the combined weight of 32 average men and women.

## Nomenclature

Somehow, I am not attracted by F. B. Rudd's suggestion in the September Wiveless W'orld that what we now know as the intermediate frequency should be styled henceforward the resultant frequency. For one thing, you could not abbreviate it to r.f., for that is already in use for radio frequency. Our frequency designations are a pretty bad muddle, anyhow. The main reason is that so many of the terms are used sometimes with a relative and sometimes with an absolute meaning. What, for instance, is a high frequency or a low irequency? Glance through disquisitions by various writers on the amplification problems in different parts of the receiving set and you will see what I am driving at. Myself, I would like to see a standardization on the following lines. Let the three main classes of frequency met with in a receiver be named respectively signal, intermediate and audio fre-quencies-s.i., i.f., and a.f.; divide each of these into three degrees:
upper, middle and lower. All the three class terms are relative, for s.f.s may be anything from a few kilocycles to thousands of megacycles, i.f.s have a smaller but similar possible range and a.f.s run from frequencies below the limits of human hearing to those a long way above those limits, but, as applied to the receiving set they are perfectly clear.

## Still They Come

There can be no doubt that television has caught on as a national hobby. By the time that this is printed the number of licences will probably have passed the halfmillion mark, which means that, despite raw material shortages and production difficulties, a very considerable proportion of the homes in which reception is possible will have television sets. Between them the London and Birmingham scrvice areas have a population of some 18 millions. Allowing an average of four persons to each household gives a figure of 4.5 million homes in the areas. For one reason or another television reception is probably impossible in at least half-a-million of these. Hence, just about one home in every eight in which reception is possible now has the necessary equipment. What will the saturation figure be? Higher costs-initial, maintenance, running and replace-ment-will naturally make it much lower than the figure for sound broadcast receivers, which are now part of almost every home. For television I predict it will turn out to be somewhere between one home iis four and one home in five.

## Terminals

The terminal (Americè: binding post) as we know it to-day and have known it for umpteen years is definitely not a satisfactory electrical device. There are only two main types. In one of these a nut clamps the wire or wires to a nonmoving base. In the other the base is again fixed, but the clamping is done by the point of a screw. The trouble in either case is that the moving part of the clamping system rotates and in so doing grinds, bruises and even cuts the wires. We're all only too familiar with defects due to the eventual breaking of wires-particularly flex-at the point of compression in the terminal. What is urgently wanted is a terminal in which a straight push, without any twisting action, is exercised on the wires.

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FREE INFORMATION SERVICE:

## CORRECTION

In our advertisement in the September issue of "Wiroless world ", it was stated in error that the now "Avo" Valve Data Manual is being supp'led tree with the "Avo" Electronic Testmeter. The Book in question is, in fact, supplied free only with the "Avo" Valve Characteristic Meter.

An important feature is the free information service, to the end of June, 1952, whereby supplementary data, to cover new valves as they are introduced, will be mailed from time to time. These supplementary items will be issued on perforated adhesive sheets so that they can be easily detached and inserted in the book in their correct alphabetical and numerical positions.

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cps, then you can, presumably, with a semitransparent, if not clear, conscience say
that the speaker has a response of 30 -
$20,000 \mathrm{cps}$. Whether this information is of any
use or not depends upon whether you want the loudspeaker to listen to or talk
about. By these standards a poor loudspeaker
looks as good on paper as a good one, which is why we try to refrain from sticking
our necks out and say as little about the performance of our speakers as possible.

When it comes to listening tests however, we find quite a lot of people choose our models, and
only a few refuse to believe their ears
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F29. Outpur, 0-2-4-5-6.3v. at 4 amps.,15)FU6 and F29 clamped with Flying Leads.F5. Output, $6 \cdot 3 \mathrm{v}$ at 10 amps . or 5 v . at 10 amps , or $12 \cdot 6 \mathrm{v}$. at5 amps. or $10 v$, at 5 amps. ................................................................
F6/4. Output, four at 6.3v. tapped at 5v. at 5 amps, per winding,
giving by suitable series and parallel connections 24v, at 5amp
20v. at }5\mathrm{ amp., 18v, at }5\textrm{amp.,}15\textrm{v}\mathrm{ , at }5\textrm{amp., 12.6v, at }10\textrm{amp
10v. at 10 amp., 6.3v, at 20 amp.. 5v, at 20 amp
31/6
F5 and F6/4 framed with Flying Leads.
OUTPUT TRANSFORMERS
MOPI. Racios 26, 46,56,66,90, 120-1 50 m/a. max. currenc.
C.T. for Q.P.P. Class B, ete. Secondary 2/4 ohms. Top panel
and clamped .............................................................................
OPI. Midget Power Pentode, ratlos 30, 60, 90-1, 40 m/a.
Secondary 2/3 ohms. .......................................each
OP2. Midget Pentode, ratios 45-1. Secondary 2/3 ohms, 40 m/a
per doz
OP10. 10/15 wates ourput. 20 racios on Full and Half primary 16/3
Op30. 30 watrs outpur, 20 ratios on Full and Half primary 12/9
Williamson's O.P. Transformer to Author's specification f3/12/6
Chokes for Williamson's Amplifier. }30\textrm{H}\mathrm{ at }20\textrm{m}/\textrm{a
Chokes for Williamson's Amplifier. 30H at 20 m/a., 15/6
10H at }150\textrm{m}/2
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Choke C.4. }60\textrm{m}/\textrm{a}.,\mathrm{ , approx. }8\textrm{H}.,350\mathrm{ ohms.
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Choke C6. }50\textrm{m}/\textrm{a}.,50 H., 1,500 ohms
Choke C6. }50\textrm{m}/\textrm{a},5,50\textrm{H}.,\textrm{l},500\mathrm{ ohms......

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        ch 18/6
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Output
FSM63. (Midget) Output $250 / 0 / 250 \mathrm{v} .60 \mathrm{~m} / \mathrm{a} .6 .3 \mathrm{v}$, at 3 amps 5 v . at 2 amps
$19 / 6$

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FS75. $275 / 0 / 275 \mathrm{v}, 100 \mathrm{~m} / \mathrm{a}$
FS $30 \mathrm{X} . \quad 300 / 0 / 300 \mathrm{v}, 100 \mathrm{~m} / \mathrm{a}$
$21 / 6$
FS3X. $350 / 0 / 350 \mathrm{v} .100 \mathrm{~m} / \mathrm{a}$. .................................................... $21 / 6$
All the above have 6-3-4-0v, at 4 amps ., $5-4-0 \mathrm{v}$, at 2 amps.
FS43. Output, $425 / 0 / 425 v .200 \mathrm{~m} / \mathrm{a} .6 \cdot 3 \mathrm{v}, 4 \mathrm{amps}$. C.I. $6 \cdot 3 \mathrm{v}$.
F550. Output. $450 / 0 / 450 \mathrm{v} .250 \mathrm{~m} / \mathrm{a} .6 .3 \mathrm{v}, 2 \mathrm{amps} . \mathrm{C} . \mathrm{T} ., 6.3 \mathrm{v}$
4 amps. C.T. 5v. 3 amps. Fully shrouded $\ldots \ldots . . . . . . . . . . . . . . . . . . .$.

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F35X. Output, $350 / 0 / 350 \mathrm{v}, 250 \mathrm{~m} / \mathrm{a} .6-3 \mathrm{v}, 6 \mathrm{amps}, 4 \mathrm{v}, 8 \mathrm{amps}$.,
FSI60X. Output $350 / 0 / 350 \mathrm{v}$. $160 \mathrm{~m} / \mathrm{a}$., 6.3 v . 6 amps, 6.3 v .

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HS6. Ousput, 251355 . Half shrouded.................................................
For receiver R1 3 amps. Half shrouded.
Half shroud, $250 / 0 / 250 \mathrm{v} .100 \mathrm{~m} / \mathrm{a} .6 \cdot 3 \mathrm{v} .6 \mathrm{amps}$. C.T. 5 v .3 amps .
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The above have inputs of $200 / 250 \mathrm{v}$,
RANSFORMERS
F4. Output, $4 v, 2 \mathrm{amps}$.
10 v , at 10 amp .6 .3 v as 20 mp . 5 v at 20 amp . 6 v . at 10 amp .,
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and clamped .............................................................................................. OPI. Midget Power Pentode, ratios $30,60,90-1,40 \mathrm{~m} / \mathrm{a}$, ,
Secondary $2 / 3$ ohms.....................................................................$~$
OP2. Midget Pentode, ratios $45-1$. Secondary $2 / 3 \mathrm{ohms} .40 \mathrm{~m} / \mathrm{a}$.
OP10. $10 / 15$ wates ourput. 20 racios on Full and Half primary 16/3
OP30. 30 wates output, 20 ratios on Full and Half primary...... 23/9
Williamson's O.P. Transformer to Author's specification $\mathbf{6 3}$
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\begin{tabular}{|c|c|c|c|c|c|c|}
\hline Type & D.C. Inpur & A.C. Outpur & Wartage Load & \multicolumn{3}{|c|}{Price} \\
\hline 230/300-A & 200/250 v. & 200/250 v. & 200/300 & ¢16 & 0 & 0 \\
\hline 230/300/110-A & 100/120 v. & 200/250 v. & 150/300 & f16 & 0 & 0 \\
\hline 230/200-A & 200/250v. & 200/250 v. & 100/200 & 614 & 0 & 0 \\
\hline 230/200/110-A & \(100 / 120 \mathrm{v}\). & 200/250 v. & 100/200 & 614 & 0 & 0 \\
\hline 230/200/50-A & 50 v . & 200/250 v. & 100/200 & ¢16 & 0 & 0 \\
\hline 230/110-A & 180-270 v. & 200/250 v. & 60/110 & 110 & 15 & 0 \\
\hline
\end{tabular}

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& to & to & to \\
& \(50 \AA\) & \(50 \AA\) & \(50 \AA\)
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\begin{tabular}{rlll}
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The line output stage of a television receiver imposes conditions which call for a valve having properties which differ considerably from those of a normal output valve.

The essential requirements for a line output valve are, first, that it shall be capable of supplying a large peak anode current when operated at a low anode voltage, and second, that the insulation of the anode shall be capable of safely withstanding the high voltage peaks, which may amount to 5 KV to 7 KV , occurring during the fly-back period:

Satisfaction of the first of these requirements is mainly a matter of the geometry of the valve; the second can be assured only by careful mechanical design.

The PL81 is an all-glass power pentode on the B9A (Noval) base, specially designed for use as a line output valve. Its heater is rated at \(0.3 \mathrm{~A}, 21.5 \mathrm{~V}\), and is therefore suitable for series operation in transformerless receivers intended for A.C. or D.C. mains supply.

It will deliver a peak anode current of 350 mA at an anode voltage of only 70 V and a screen voltage of 170 V . The efficiency of the yalve is thus intrinsically high. By adopting the booster diode circuit whereby the H.T. line voltage is increased, the required output can be obtained at a still lower anode voltage, so that the already high efficiency of the stage is further improved.
The necessary high insulation resistance to withstand a peak voltage of 7 KV has been obtained by special internal construction, and by bringing out the anode connection to a top cap so that the use of a specially insulated valve-holder is not necessary. Even so, the seated height of the valve is only 75 mm and the overall diameter 22.2 mm -dimensions only slightly larger than those of the voltage amplifying pentode in the same range. The production of a valve of this output and high insulation in so small an envelope may be counted as something of an achievement.

\section*{RATINGS \& CHARACTERISTICS}
\begin{tabular}{|c|c|}
\hline \multicolumn{2}{|l|}{Heater Suitable for Series operation, A.C. or D.C.} \\
\hline \(V_{h}\) & 21.5 V \\
\hline In & 0.3A \\
\hline \multicolumn{2}{|l|}{Capacitances} \\
\hline \(c_{\text {in }}\) & 14.3 \({ }^{\mu} \mu \mathrm{F}\) \\
\hline cout & \(6.5 \mu \mu \mathrm{~F}\) \\
\hline \(\mathrm{C}_{\mathrm{a}-\mathrm{gl}}\) & \(<0.4 \mu \mu \mathrm{~F}\) \\
\hline \(\mathrm{c}_{\mathrm{g}} \mathrm{l}\)-h & \(<0.2 \mu \mu \mathrm{~F}\) \\
\hline \multicolumn{2}{|l|}{Characteristics} \\
\hline \(V_{\text {a }}\) & 170 V \\
\hline \(V_{83}\) & OV \\
\hline \(V_{\mathrm{g} 2}\) & I70V \\
\hline \(\mathrm{V}_{\mathrm{g}}\) & -22V \\
\hline \(\mathrm{Ia}_{1}\) & 45 mA \\
\hline 1 g 2 & 3 mA \\
\hline gm & \(6.5 \mathrm{~mA} / \mathrm{V}\) \\
\hline \(\mu_{\mathrm{g}}^{1-\mathrm{g} 2}\) & 6 \\
\hline
\end{tabular}

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MINIATURE AMERICAN TYPES for AC/DC EQUIPMENT

A complete range of miniature B7G valves is now available for 110 volt or 240 volt AC/DC equipment. All these valves bear American type numbers.

TYPE 50C5
OUTPUT TETRODE for 110 v . Supply
Heater Rating
50.0 volts 0.15 amp

Mutual Conductance
Power Output
\(7.5 \mathrm{~mA} / \mathrm{V}\)
1.9 watts

Price: 15/6 plus P. Tax

TYPE 19AQ5
OUTPUT TETRODE for 240 v . Supply
\begin{tabular}{lcccc} 
Heater Rating ... & \(\ldots\) & \(\ldots\) & \(\ldots\) & 19 volts 0.15 amp \\
Mutual Conductance & \(\ldots\) & \(\ldots\) & \(4.1 \mathrm{~mA} / \mathrm{V}\) \\
Power Output ... & \(\ldots\) & \(\ldots\) & \(\ldots\) & 4.5 watts
\end{tabular}

Price: 13/- plus P. Tax

TYPE 35W4
HALF WAVE RECTIFIER


DIMENSIONS :-Seated Height \(27^{\prime \prime}\) max. Diameter \(\frac{1}{\prime \prime}^{\prime \prime}\) maz.

TYPE I2BE6
SPECIAL HEPTODE


TYPE I2BA6
VARI Mu R.F. PENTODE
\begin{tabular}{llllll} 
Heater Rating & \(\ldots\) & \(\ldots\) & \(\ldots\) & \(\ldots\) & 12.6 \\
volts & 0.15 mmp. \\
A node Impedance & \(\ldots\) & \(\ldots\) & \(\ldots\) & 0.25 & 1.5 meg. \\
Mutual Conductance &.. & \(\ldots\) & \(\ldots\) & 4.3 & \(4.4 \mathrm{~mA} / \mathrm{V}\)
\end{tabular}

Price: 13/- plus P. Tax

TYPE I2AT6

\section*{DOUBLE DIODE TRIODE}
\begin{tabular}{lllllll} 
Heater Rating & ... & \(\ldots\) & \(\ldots\) & \(\ldots\) & 12.6 & volts 0.15 amp. \\
Voltage Gain & \(\ldots\) & \(\ldots\) & \(\ldots\) & \(\ldots\) & 33 & 42
\end{tabular}

Price : 12/- plus P. Tax

DIMENSIONS:-Seated Height \(1 \mathbf{H}^{7 /}\) man. Diameter f' \(^{*}\) max.


\section*{AN APOLDGY}

The inconvenience and annoyance caused to the public and to our friends in the trade occasioned by the delay in delivery of acos microcell pick-ups is a matter that causes us the gravest concern. It is in the main due to the unprecedented demand for these pickups which has outrun the supply of certain raw materials and also made our factory space inadequate. Commitments to manufacturers made prior to the general announcement of the pick-ups have still further aggravated the position.

Happily, the raw material problem should shortly be solved and plans for greater production facilities are being urged forward. In the meantime, we are doing all we can to meet existing orders, but regret that delivery on new orders will be subject to about three months' delay.

We can only express our regrets for the circumstances that have made this announcement necessary, and trust that our trade friends and the general public will bear with us in view of our assurance that all possible steps are being taken to meet the situation and to reduce delay to a minimum.


\section*{THE "BELLING-LEE" PACE}

\section*{Providing technical information, service and advice in relation to our products and the suppression of electrical interference}

Apologies to the B.B.C.
In the September issue of the Wireless World " on this page we stated that the B.B.C. altered the frequencies of Sutton Coldfield a fortnight before the station opened.

This was a mis-statement for which we do not hesitate to apologise. Something happened, but at this distant date we cannot recall the details, but B.B.C. Engineering Division tell us that they first published the Sutton Coldfield frequencies on 20th September 1948 and that they were never altered. We agree that nothing should be said or written which might cause uneasiness in the minds of the trade or the public.

While we do our best to be accurate in this page, when we do " drop a brick," the quality of those who come back at us is ample evidence that our page is read by the right people.

\section*{Rationalisation of Television Aerial Numbering}

Most " Wireless World " readers know that eventually there will be eleven television transmitters operating on five frequency channels, and therefore the sooner we stop talking about the London and Midland transmitters the better. The transmitters should be referred to by their official channel number -London-Channel I, Midland-Channel 4, etc.

All aerial manufacturing members of the Radio and Electronic Component Manufacturers' Federation have agreed to indicate the television frequency by the official channel number following a stroke, incorporated immediately after the list number of the particular aerial.

It Jas also been agreed to standardise a colour code for each frequency chanuel, but this may not be used by everyone immediately.

So far as all this concerns " Belling-Lee" a complete new list of television aerial numbers has been built up, and became operative on \({ }^{\circ}\) Sept ist.

At first all aerials are being despatched from the factory labelled on the outside with the old and the new numbers, and using the following colour code:
Channel No. Cat. \begin{tabular}{c} 
Colour \\
suffix \\
Code
\end{tabular}
\begin{tabular}{c} 
I. London \\
2. North England \\
2.
\end{tabular} Lellow
3. Light Blue
the first suffix the channel on which the aerial will operate (-/I London, -/4 Midland, etc.) and the second suffix shows either the type of mounting (-/L. chimney lashing, - - W wall mounting, -/9 9 foot mast and lashing) or the type of feeder supplied ( \(-/ \mathrm{B}\) balanced, \(-/ \mathrm{U}\) unbalanced).


L802/1 or 4/C. L802/1 or 4/L. Comparison of Aerial Prices
"Belling-Lee" aerials may appear dearer than some but we do ask those interested to make certain that when they are thinking or talking aerial prices that they are clear in their minds what they are getting for their money. "Belling-Lee" have made a practice of stating the price of a complete aerial, whereas some other manufacturers feature a price which is in fact the price of the cross-arm and elements only, pole, if any, and lashings are extra, Because of this, one is apt to consider that "Belling-Lee" aerials are more expensive than they really are.


It is very gratifying that some interest is at last being taken in this subject and it is possible that a lot will be done voluntarily before legislation takes place. The A.A. give it its blessing, then the B.B.C. are helping by their film on Television and by demonstration on Sound Broadcast, neither are they always preaching to the already converted. There must be many television viewers whose cars are still a source of annoyance to others. The selfish owner never sees the trouble his own car makes, unless his wife drives too, and then it is only for a minute while the car is put in its garage. It is the duty of every viewer to suppress his own car and to persuade all those who see his television to do likewise Many garages have been slightly hostile towards the fitting of suppressors, so we are grateful to the
" Motor" for the excellent article they published recently. This, together with the fact that one of the greatest names in car electrics now recommend and market suppressors, must do a great deal to break down uninformed opposition. An Amateur testifies on behalf of the "Coronette" silent aerial discharger.


This is an extract from a letter dated July:
" The past week or so has given ample opportunilies for studying the effects of the device on coronary discharge' interference with the reception of short and medium wave broadcast reception.

I am using an R.C.A. 'AR.77E Communication Receiver and can now listen in comfort to the most distant continental medium wave transmitters during heavy and protracted rain storms. It is only after half an hour's torrential rain that a slight background 'hiss' manifests itself, and this is a remarkable improvement after my experience before fitting the Discharger, when, as you know, immediately rain started any programme, other than North Regional or Light, was blotted out by terrific 'static' of the shrieking variety.

I am happy to say that the expense incurred in taking down and re-erecting the aerial consequent upon fitting the 'Silent Discharger' has been more than repaid in increased listening pleasure over the last two week-ends.

I might add that I seem to get all-round quieter performances since fitting your Company's device, and am very grateful to you for all the trouble and interest you have taken in the matter."
NOTE: The "Coronette" discharger is designed to fit " Skyrod " aerials only and is unsuitable for other makes of Vertical aerials.

\section*{An amAZing new amplifier.}

\section*{Jfactual jfigures}

FROM A-Z, FROM PICK-UP TO SPEAKER, including Tone Control Circuits NOT MERELY A LABORATORY MEASUREMENT OF AN AMPLIFIER ONLY, BUT THE WHOLE STORY OF THE : A-Z", AS IT WILL PERFORM UNDER USERS' CONDITIONS -NATIONAL PHYSICAL LABORATORY REPORT
ON TESTS OF AMPLIFIER (MARKED : TYPE A-Z, SERIAL No. A50/I, SOUND SALES) TESTED FOR :-SOUND SALES LTD., WEST STREET, FARNHAM, SURREY REFERENCE:-LETTER, RNW/DES/AZ, DATED 31.8 .50
The instrument was tested for sensitivity at various frequencies between 20 and 25,000 cycles per second. The total harmonic distortion was measured at \(60 \mathrm{c} / \mathrm{s}\) and \(1,000 \mathrm{c} / \mathrm{s}\); the hum and noise output voltage and the output impedance were also determined.
The supply to the instrument was at 230 volts and \(50 \mathrm{c} / \mathrm{s}\); the connection to earth was independent of the mains supply and made directly to the instrument.
The input was in all cases made to the lead provided, with the controls set as follows:-LF cut, 0 ; volume 12 (maximum) ; Bass, 0 ; Treble, 0. A load resistance of 15.0 ohms was connected to the output of the amplifier by means of the leads provided.

TABLE I
8 WATTS OUTPUT

\section*{SENSITIVITY}

The input required to develop an output of 12 watts in the load at \(1.000 \mathrm{c} / \mathrm{s}\) was \(0.125 \pm\) 0.002 volc.

The input required to develop an output of 8 watts was measured at several frequencies the results obtained are given in Table i

\section*{HARMONIC DISTORTION}

The harmonic distortion was measured with an output of 12 watts at \(60 \mathrm{c} / \mathrm{s}\) and \(1,000 \mathrm{c} / \mathrm{s}\). The second and third harmonics of the fundamental, in about equal proportions, were found to predominate. The total harmonic voltage expressed as a fraction of the fundamental is given in Table II.

\section*{HUM AND NOISE}

The hum and noise output voltage was measured with the input lead short circuited and found to be 0.005 volt.

\section*{OUTPUT RESISTANCE}

The output resistance was measured at a The output resistance was measured at
Irequency of \(1,000 \mathrm{c} / \mathrm{s}\) for output powers of 3 watts and 12 watts in a load resistance of 15.0 ohms.

The resultes obtained are given in Table III
Date 5th October 1950.
Reference E.488.18.
Passed by L.M.F. C.H.R.
\begin{tabular}{|c|c|}
\hline Frequency, e/s & Input, volt \\
\hline 20 & 0.100 \\
60 & 0.101 \\
1,000 & 0.102 \\
3,000 & 0.102 \\
5,000 & 0.102 \\
10,000 & 0.102 \\
15000 & 0.102 \\
20,000 & 0.102 \\
25,000 & \\
\hline
\end{tabular}

TABLE II
12 WATTS OUTPUT


TABLE III
\begin{tabular}{|c|c|}
\hline Power Output, watts & Output resistance, ohm \\
\hline 3 & \(0.50 \pm 0.05\) \\
12 & \(1.2 \pm 0.1\) \\
\hline
\end{tabular}
- E. C. BULLARD, Director.
L. HARTSHORN,
for Superintendent, Electricity Division.

YOU ARE WELCOME TO A DEMONSTRATION OF THIS AMAZING NEW AMPLIFIER, WHICH COSTS ONLY \(£ 30\) COMPLETE, AT OUR LONDON OFFICE \& SHOWROOMS.

\section*{Found Males Limited}

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\section*{Mk. VII}

The Stereophonic Amplifier Mk. VII, according to users, gives the highest quality of any reproducer so far produced. This high quality is to be expected when the particular points of the design are considered.
The large reduction of intermodulatiou distortion and Dopler effect is achieved by completely separate amplifier channels for the high and low frequencies, terminating in separate speakers. The latter may be separated by any convenient distance to prevent pressure wave modulation and it is unnecessary to restrict the bass to prevent marring treble response.
In each case the speaker is damped by a minimum of 12 times at audible frequencies to prevent speaker resonance, and the crossover is very fast, approximately 28 db for the first octave. A control is provided to vary the power applied to the treble speaker, to compensate for its efficiency compared with the bass speaker, and a frequency record reveals the absence of dip or rise in output at crossover frequencies.
The treble control gives 20 db lift or 14 db cut from 3,000 cycles and compensates for loug-playing records and various American and English recording characteristics.
The small improvements in the latest model amplifier result in reproduction which is extremely good and capable of pleasing all tastes.

Chassis complete with valves
\[
\text { Price } 36 \frac{1}{2} \text { gns }
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 you receive is one in ten million speakers which are proving the superiority of Rola performance throughout the world. So let figures speak for themselves and specify Rola.

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SOUTH:-
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It is impossible to detail even briefly in this small space the exhaustive ground covered by this comprehensive work, but the pamphlet we can send you will show you that these few remarks are no exaggeration.

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\section*{REMOTE CONTROL PRE-AMPLIFIER \\  \\ \\ £8.15.0 \\ \\ £8.15.0 For use only with LEAK For use only with LEAK "POINT ONE" AMPLIFIERS} "POINT ONE" AMPLIFIERS}

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3. 78 r.p.m. Records, with built-in equalisation for British characteristics.
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5. Radio (or other sourct). Response, level. Sensitivity, 50 mV r.m.s. input resistance, \(100,000 \Omega\).

Treble Control. A seven-position switch allows the choice of accurately determined boosts and losses on all inputs.
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Panel Light as a visual reminder.
Total Harmonic Distortion : 0.02\% approx.
The sensitivities relate to 10 watts output at \(1,000 \mathrm{c} / \mathrm{s}\), with the tone-controls level, when coupled to the TL//2 power amplifier which provides the heater and anode supplies.

These amplifiers are those used by H. J. LEAK when giving the most recent of his demonstrations between a reproduced orchestra and (two minutes later) the live orchestra in the lecture theatre.
The amplifiers can be used for disk recording with the assurance that total amplifier distortion will certainly be no greater than that obtaining in the major studies of the world.

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These units mearure \(2 \times 1 \times 7 \mathrm{in}\). and provide th \({ }^{e}\) solution to obtaining a matehbetween low impedane? phones and high impedance output.
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\text { Valves, etc, Post pald. } 84=2=6
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10in, Figh Quality Speaker by tamous manufacturer at the special price of \(15 / 11\) (Note: this offer only applien to those purchasing the complete amplifier kit).

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\section*{M.O:S "FULVISION"}

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5 KV. ELECTROSTATIC VOLTMETER. Seale length 31 in ., fush mounting, 41 in. diameter, £2/10/-. BATTERY CHARGERS
Input 100/220 F . A.C. Output 15 volts at 16 amps. Cont tnuously variable metered output. Usual price \$24. Our price, \(210 / 10 /\)-each, plus 10 /-carriage.
SPECLAL HEADPEONE OFFER. Hizh-grade Double Geadphoner, using balanced armature unith. D.C. Res 8/6 each.
SPECIAL OFFER OF ELECTROLYTID CONDENSERS \(16+16 \mathrm{mf} .500 \mathrm{v}\). working
\(8+8 \mathrm{mf}, 500\).
\(8+8 \mathrm{mf}\). 800 V. working
\(32+32 \mathrm{mf} .330\). working, all cans
38 mf .350 v . working
32 mt .350 v . Working
16 mf .450 v . worktorg, cardboard
8 mf . 430 v . workhog, cardboard
4 mt .500 v . worktng, cardboard
\(16+8450 \mathrm{v}\). working, all. cans
All Capacities and Voitages a ralinble.
LOUDSPEAKERS by famous makers.
\[
\begin{array}{lr}
31 \mathrm{in} & 9 /- \\
\sin . & 10 / \text { in. } 12 / 6 \\
\sin . & 13 / 6
\end{array}
\]

WILLIAMSON AMPLIFIER KIT.
We can supply the Kit of Parts for the latent veralon of this famous amplifier complete in every detall for
flo/ \(10 /=\), with Valvea.

WILLIAMSON AMPLIFIER OUTPUT TRANSFORMERS
to epecificatlon, 63/-.
Matns Transformers, \(45 /-1\)
Hit. ELIMINATOR AND TRICKLE CHARGER EIT. All parts to construct an eliminator to give an output of 120 volts at 20 man . and 2 volts to charge an accum ulator. Uses metal rectifler, 35/-
\(4 / 11\)
\(4 / 3\) 411
216


4/11

ORAMOPHONE AMPLIFIER KIT
Consints of Complete Klt of Parts for a 2| watt, Mainsoperated 2 -stage Amplifler for uae with any type of pick-up. Volume and tone controls are incorporated Outputimpedance is 8 ohms. Cat. No. Ampla. Price complete, 65/-. Vor \(200-230 \mathrm{y}\) MAINS NOISE ELIMINATOR
MAINS NOISE ELIMINATOR KIT.
Tuo specially denigned chokes with tharee smoothing condensern with clruit diagram. Cuts out all malng
noine. Can be assembled lngide existing recelver. fycomplete.
CO-AXIAL CABLE.
Super quality cable, contisting of a centre copper core a polyviniresin type insulator, s flexiblescreen, a weather-
proof P.V.C. outcr cover. Just the thing for Television lead-in, super mike cable, etc., 80 olums linpedance. Cat. No. C.755, 3d. per foot.

\section*{PREMIER 1950 CATALOGUE}
contains all the newest TV Kits, Components, Aerials, Tubes, etc., in addition şo thousands of Radio Bargains. Price-6d.
T.V. WHITE RUBBER MASKS.

We can now supply a apecially designed White Rubber Mask for 6 in . C.R. Tubea at \(7 / 6\) each.
gin. White Masks, 9/6. 12in. White Maska, 15/0.
SUPER QUALITY TELEVISION MAGNTFYING LENS To suit \(5 \ln\)., 6in. or 7 m . Tubes. Increase pleture size considerably, 25/-each.
PERMANENF MAGNET FOCUS POTS. Avajlable for all Tubes, 15/-. Please state Tube used.
MOVING COIL METERS. All 24 m, outalde dameter mA. \(7 / 18\). 5 mA 5/. 50 mA , 816 . 180 mA . 20 : เmp., \(7 / 6\); 40 amp. \(7 / 6\); \(20 \mathrm{v.,5/9}\); 40 v., 5.9 ; 500 microsmpr, \(7 / 8\). All \(3!\mathrm{in}\). outaide diameter. 1 mA . 15/11: \(30 \mathrm{~mA}, 10 / 6\); \(200 \mathrm{~mA}, 8 / 6 ; 500\) microampa, 19/6; Thermocouple meters, \(21 \mathrm{in} ., 2.5 \mathrm{amp} ., 5 /-;\), 3 static 3 tin. \(, 2 \mathrm{kV},, 25 /=\)

THE LAST FEW R 107 RECEIVERS. ONE OF TEE ARMY'S FINEST COMMUNICATIONS



9 valves, R.F. amp. osc. Frequency Changer, 2 1.F.s. (465 kc.), 2nd betector, A.V.C. Af. amp. B.F.O. A.C. mains, \(100-250\)
17.5 to \(7 \mathrm{Mc} / \mathrm{s}, 7.35 \mathrm{Mc} / \mathrm{g}\) to \(2.9 \mathrm{Me} / \mathrm{s}, 2.0\) to \(1.2 \mathrm{Mc} / \mathrm{s}\). Monitor L. S. bullt in. Complete. Writefor full detalle. Price 212/12/-, plus 21/-carriage and packing.
A.C. ALL-WAVE SUPERHET KIT

7 valven (plus rect ifters) for \(200-250\) v., \(40-60\) cycle A.C.
 metres.
Ptck-up Input. Uses 6K7, 6K8, 6K7, 6B8, 6J5 and 2.6 V 6 in pusb-pull, glving an output of 10 watte. 3 and 15 ohm speaker
Negative ieedbatk is applied over 3 atages giving a high fldelity output. Tone control is imeorporated. £14/5/-. Also available for AC. D.C. Sains. Spectication as \(6 \mathrm{~J} 7,2\) - K T33C, In Kitform at \(213 / 8 / 10\)

\section*{FURTHER EXPANSION! ALL POST ORDERS SHOULD NOW BE SENT TO OUR NEW HEAD OFFICE AT 740 HIGH RD., TOTTENHAM, N. 17. Telephone: TOTTENHAM 5371.}
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|c|}
\hline  &  & \multicolumn{3}{|l|}{} &  &  &  &  &  & &  \\
\hline £4.19,6 me. P. & & \multicolumn{10}{|c|}{transmitting and special purpose valves} \\
\hline  &  &  & TRA &  &  &  &  &  & & & \\
\hline
\end{tabular}

152-153, FLEET STREET, E.C.4. Phone: CENtral 2833-and at-207 EDGWARE RD., W.2. Phone: AMBassador 4033 207. Edgware Road is open untll 6 p.m. on Soturdays, Terms of Business: Cash with order or C.O.D. over \(\mathcal{C l}\).

\title{
DREMITE MADD COMIMTY NEW LONG RANGE TELEVISOR KITS
}

FOR THE LONDON OR BIRMINGHAM FREQUENCIES
£17-17-0

As is usual in Premier Kits every single item down to the last Bolt and Nut is supplied. All chassis are punched and layout diagrams and theoretical circuits are included.

FIVE EASY-TO-ASSEMBLE KITS ARE SUPPLIED :-


A well-made walnus finish PEDESTAL CABINET is available from stock ae € \(\$ / 1010\), plus \(7 / 6\) carriage and parking. Working models can be seen during transmitting hours at our Fleet Street and Edgware Road branches.

VISION RECEIVER with valves, carriage \(2 / 6 \ldots . . . . . . .\). E3 136 SOUND RECEIVER with valves, carriage \(2 / 6\).......... \(£ 2146\)
TIME BASE with valves, carriage 2/6........................ \(£ 276\)
POWER SUPPLY UNIT with valves, carriage \(5 / \ldots \ldots\) \& 6 ,
TUBE ASSEMBLY, carriage and packing \(2 / 6 \ldots \ldots \ldots . .\). ...... \(£ 2\) I8 6
This unit includes the VCR97 Tube, Tube Fittings and Socket and a 6 in . P.M. Moving Coil Speaker with closed field for Television. The Instruction Book costs \(2 / 6\), but is credited if a Kit or the complete Televisor is purchased.
Any of these Kits may be purchased separacely ; in fact, any single part can be supplied. A complete priced list of all parts will be found in the Instruction Book.
20 Valves are used, the coils are all' wound and every part is tested. All you need to build a complete Television Receiver is a screwdriver, a pair of pliers, a soldering iron and the ability to read a theoretical diagram.

The rollowing sensitivity figures prove that the Premier Televisor Kit is capable of reception at greater distances than any other standard commercial ki. or receiver whether T.R.F. or Superhes.

> VISION RECEIVER.


\section*{NEW PRE-AMPLIFIER FOR FRINGE RECEPTION AREAS With sensitivity better than 35 db .}


We can supply the complete kit of parts to make this wide band width Pre Amplifier, using 2 717A Pentodes. With all parts, valves, chassis, diagrams, etc., 27/6
All parts available separately.

\section*{SEPARATE POWER SUPPLY KIT}
for use with this Pre.Amplifier can be supplied for 21/=
(H.T. output 130 v ., 20 mA . L.T. 6.3 v ., . 4 A .

When ordering Televisor kits
PLEASE STATE IF THE LONDON OR BIRMINGHAM MODEL IS REQUIRED

\footnotetext{
207, EDGWARERD:, W. 2 Phone: AMBassador 4033
(Open till 6 p.m. Satur AND AT 152-153, FLEET STREET, E=C.4 Phone: CENtral 2833

Terms of Business : Cash with order or C.O.D. over fl. Send 6d. Stamp for list.
ALL POST ORDERS to our new Head Office: 740 HIGH ROAD, TOTTENHAM, N.17. Phone : TOTtenham 5371.
}

\section*{EDREMIER PADIO COMMAMY}

\title{
Now being Delivered! PREMIER TELEVISOR KITS
} FOR LONDON AND BIRMINGHAM USING 9" OR 12" MAGNETIC C.R. TUBES

(Carriage and Packing 15/-) including all parts, valves and loudspeaker, but excluding C.R. TUBE

\section*{CIRCUIT DETAILS}

The Visipn Receiver consists of 4 R.F. stages (EF54's) which are followed by a Diode Detector and Noise Limiter (6H6) which is directly coupled to the Video valve (EF54). Complete Kit with valves, \(\mathbf{6} / \mathbf{1 6} / \mathbf{0}\).
Carriage and packing 2/6.
The Sound Receiver comprises 3 R.F. stages ( \(65 \mathrm{SH}^{\prime}\) 's) followed by a Double Diode Triode (6Q7), which acts as Detector and L.F. Amplifier. A Noise Limiter (EA50) is also incorporated. The output valve ( 6 V 6 ) drives a 10 in . P.M. Moving Coil Speaker with closed field magnet, which is included In the Time Base Kit.
Complete Kit with valves, \(\mathbf{6} / 1 / 0\).
Carriage and packing 2/6.
The Time Bases employ blocking oscillators on both Line (6SH7 and 807), and Frame (VRI37 and 6V6). E.H.T. (Non-lethal) is taken from the Line Output Transformer through a voltage doubler employing two valves (VUIII). The Sync separators are 6H6 and 6 V 6 .
Permanent Magnet Focusing.
Complete Kit with valves, \(\mathbf{\varepsilon 8 / 5 / 6}\).
Carriage and packing \(5 /-\).
The Power Supply is from a double wound mains transformer completely isolating the receiver from the malns. The H.T. Rectifier is a \(5 \cup 4 \mathrm{G}\).
Complete KIt with valves \(\epsilon 4 / 16 / 6\) Carriage and packing 5/-.


The sollowing zensitivity figures prove that the Premier Televisor Kit is capable of reception at greater distances than any other standard commercial kit or receiver whether T.R.F. or Superhet.
VISION RECEIVER

Sensitivity : \(25 \mu \mathrm{~F}\) for 15 V peak to peak measured at the Anode of the Video \(V\) alve
Sound Rejection : Better than 40 db .
Adjacent Sound Rejection: Midland Model. Betzer than 50 db

\section*{SOUND RECEIVER}

Sansitivity: \(20 \mu \mathrm{~V}\). Vision Rejection: Bztcer than 50 db .

ALL POST ORDERS to our new Head Office : 740 HIGH ROAD, TOTTENHAM, N. 17. Phone : TOTtenham 5371.

\title{
DON'T MISS THESE BARGAINS
}

\section*{MAINS TRANSFORMERS}

\section*{Ordinary-E.H.T.-Special.}

All transformers are suitable for \(200-250\) v. 50 cycle mains, are fitted with primary screens, fully impregnated and complete with all necessary clamps, etc.

300-0-300 at 80 mA .6 .3 v . at 4 amp .5 v. at 3 amp . Half shrouded drop through, \(14 / 6\).
\(260-0-260\) at 60 mA .6 .3 v . at 3 amp .5 v . at 2 amps . Half shrouded drop through. Price 13/9.
350-0-350 at \(80 \mathrm{~mA} ., 4 \mathrm{v}\). at 6 amps. 4 v. at 3 amps., drop through or upright mounting, a useful replacement type for old-type receivers. 16/6.

350-0-350 at \(100 \mathrm{~mA} ., 6.3 \mathrm{v}\). at 5 amps., 5 v. at 2 amps. Fully shrouded upright mounting, 19/6.
260-0-260 at 60 mA .4 v . at 4 amps. 4 v. at 3 amps. Half shrouded drop through. Price 13/9.


235-0-235 at 60 mA . 6.3 v . at 3 amp ., 5 v . at 2 amps . Fully shrouded upright mounting, \(15 \%\). Primary has an additional tap for 110 v .

Filament Transformer, 6.3 v . \(1 \frac{1}{\mathbf{3}-2} \mathrm{amps}\), upright, \(6 /-\).
350-0-350 at 160 mA ., 5 v . at 3 amps., \(6.3 \mathrm{v}{ }^{-1}{ }^{\text {at }} 6 \mathrm{amps} ., 6.3 \mathrm{v}\). at 3 amps. Fully shrouded upright mounting, 36/-.
\(250-0-250\) at 100 mA ., 5 v . at \(3 \mathrm{amps} ., 6.3 \mathrm{v}\). at \(6 \mathrm{amps} ., 27 / 6\).

2,500 v. at \(5 \mathrm{~mA} ., 4 \mathrm{v}\). at 1.5 amp., 202 v. at \(2 \mathrm{amp} ., 27 / 6\).
4 KV . at \(5 \mathrm{~mA} ., 2 \mathrm{v}\). at 2 amp ., potted with insulators, 67/6.

Special combined transformer designed to be suitable for Televisor or oscilloscope using a 6 in . tube VCR97, etc. \(400-0-400\) at 150 mA ., 2.5 KV . at 5 mA ., 6.3 v. at \(6 \mathrm{amps} ., 6.3\). v. at 6 amp., 6.3 v . at 2 amp ., 4 v . at \(1 \mathrm{amp} ., 4 \mathrm{v}\). at \(2 \mathrm{amps} ., 5 \mathrm{v}\). at 3 amp., 70/..

\section*{SPECIAL}
6.3 v. at 2 amps.,

60 v . at 200 mA .
tapped 200/220, 230/250, 7/6

PRE-AMPLIFIER


The above is the circuit diagram of pre-amplifier model K . You will notice that the circuit includes two double valves, push-pull input, and push-pull output transformers, a relay, a choke, and numerous small resistors and condensers. The whole thing is complete with valves wired up and enclosed in a metal case size approximately \(5 \mathbb{i} \mathrm{in} . \times\) 4in. \(\times 4\) in. You will readily appreciate that this little unit will fit in almost any cabinet, including table models. These units are very wellmade by a famous American company, originally intended for Forces use, to increase the output of low gain microphones and pick-ups. We have a fair quantity available, all new and unused, the price is 17/6 each and as this is the equivalent of a 4 -valve push-pull amplifier, it is of course enormous value. We anticipate a big demand, so please order by return.


\section*{PARTS FOR TELEVISORS}


\section*{FRAME AND LINE COIL ASSEMBLY} Perfectly made by a very famous maker, for standard type magnetic tubes, 9 in ., 10 in ., 12 in ., or 15 in ., we have a limited number only, the price is 16/6, and cannot be repeated once these are cleared, so please act quickly.

PERMANENT MAGNETIC FOCUSING UNITS. No current drain-for all makes of tubes-patented method of adjusting the gap, giving really clean pictures and even focus of whole of the tube free. Price \(16 / 6\) each.

R.F. E.H.T. We were so pleased with the quality of this unit that we have taken up the entire output of the manufacturer, the voltage can be adjusted to make it suitable for working 9 in ., 12 in . or 15 in . tubes, the unique design and vacuum impregnation combine to give a trouble-free unit which will give years of satisfactory service, and, of course, the big point about the R.F. E.H.T. is that it is not lethal, the size is only \(4 \frac{1}{4} \times 3 \frac{4}{4} \times 4 i_{n}\), price complete, ready to operate, \(65 / \mathrm{m}\).

HOUSE TELEPHONES


Desk or Wall Mounting
Suitable for intercommunication between offices, workshops, stores, garages, big houses, kitchens, etc. Each station consists of normal size Bakelite handsets and desk or wall mounting cabinet with built-in selector switch, buzzer and push. All stations can communicate with one another independently. Each installation is absolutely complete and internally wired. 3 -station installation ally wired. 3 -station installation complete with 50 yards 5 -core cable \(66 / 10\)-. 4 -station installation, complete with 50 yards 6 -core cable, \(\mathrm{f} 8 / \mathbf{1 0} /=\). 2 -station installation, complete with 2 yards 4 -core cable, f3/17/6.


ELECTRIC HEATERS
Heavy cast framework totally encloses the elements the elements, so
safe even in confined spaces, just right for your radio den, garage, office, shop, etc. 900 watt (heavy) model, \(23 / 6\), plus 5/6d, carriage. Other models availmodels avaiable are: 750
watt, 500 watt,
 250 watt, all these are \(23 / 6\) each, plus \(4 / 6\) each carriage. The 250 watt model used as a foot-warmer keeps legs and body warm for less than a farthing per hour.


\section*{SHORT TURN'S COIL} TESTER

You know that it is almost an impossibility to test for shorted turns in I.F. Transformers, Coils, L.T. Transformers, etc, with an ordinary ohmmeter. Our mainsoperated shorted turns coil tester will reveal these faults in a second. An essential instrument in all coil-winding shops as it will test for continuity at the same speed. £6/10/-.

\section*{MISCELI_ANEOUS ITEMS}

8 kv . sleeving, \(3 /-\mathrm{doz}\). yds. B7G ceramic valve holder, rod. Ditto, EF50, 9d. Pax EF50, 6d. Int. Oct. Valve holder Amphenol, 6d. Pax, 4d. Miniature 100 K pots, \(1 / 6\).

DEPENDABLE SPARES. Unless otherwise stated none of our spares is Government Surplus. Prices do not include postage, for details of this see bottom of this page.

\section*{ELECTROLYTIC CONDENSERS}
\begin{tabular}{|c|c|c|c|}
\hline 4 mfd .450 v. & & 8 & 3 \\
\hline 8 mfd .150 v & & 8 mfd . \(\times 16 \mathrm{mfd}\) at 350 v . & 26 \\
\hline 8 mfd .350 & 16 & \(8 \mathrm{mfd} \times 16 \mathrm{mfd}\). at \(450 \nu\). & 36 \\
\hline 8 mfd .450 & 111 & \(16 \mathrm{mfd} \times 16 \mathrm{mfd}\) at 350 v & 3 \\
\hline 8 mfd . 500 & 26 & \(16 \mathrm{mfd} . \times 16 \mathrm{mfd}\). at 450 v. ... & 3 \\
\hline 16 mfd .450 v . & & \(16 \mathrm{mfd} . \times 8 \mathrm{mfd} . \times 24 \mathrm{mfd}\). at & \\
\hline 16 mfd .450 v . & 28 & 450 v & 42 \\
\hline 16 mfd .500 v . & 36 & \(25 \mathrm{mfd} . \times 25 \mathrm{mfd}\). at 200 v . & 311 \\
\hline 32 mfd .350 v . & 28 & \(16 \mathrm{mfd} . \times 8 \mathrm{mfd}\). at 350 v . & 2 \\
\hline 32 mfd .450 v . & 3 & \(16 \mathrm{mfd} . \times 8 \mathrm{mfd}\) at 500 v . & 4 \\
\hline 10 mfd .25 & 10 & 250 mfd . at 12 v & \\
\hline 25 mfd .25 & 10 & \(16 \mathrm{mfd} \times 32 \mathrm{mfd}\). at 350 v & \\
\hline 50 m & 10 & 12 mfd . at 50 & \\
\hline
\end{tabular}

\section*{P.M. SPEAKERS}

All speakers are by very famous makers such as Rola, Celestion, Goodmans, etc.


\section*{ENERGISED SPEAKERS}
\(6 \frac{1}{2} \mathrm{in}\). with 700 ohm field and hum buckler, price \(9 / \mathrm{-}\). 8 in . with 2,000 ohm field, hum buckler and output transformer, \(15 / 6\).

\section*{I.F. TRANSFORMERS}
\(465 \mathrm{kc} / \mathrm{s}\), iron dust cored very high " Q " fitted in standard size can, \(6 / 9\) per pair, \(465 \mathrm{kc} / \mathrm{s}\) iron dust cored very high " \(Q\) " fitted in midget size can, \(12 / 6\) per pair. \(465 \mathrm{kc} / \mathrm{s}\) air cored, medium sizc can, \(6 / 9\) per pair.

\section*{L.F. CHOKES}

Heavy duty types suitable for power packs and mains smoothing. These are mostly Government Surplus.
\begin{tabular}{|c|c|c|c|c|c|}
\hline 200 mA .10 & 10 & & 120 mA .9 henry & & \\
\hline 200 mA .3 henry ......... & 5 & 0 & 80 mA . 10 henry & & \\
\hline Midget-type inductance & & & 80 mA .5 heniry & & \\
\hline unknown & 3 & 6 & 40 mA .5 henry & 3 & \\
\hline
\end{tabular}

\section*{OUTPUT TRANSFORMERS}


\section*{VOLUME CONTROLS}

All have full-length spindle and are complete with fixing nuts. We stock full range of values between 2,000 ohms and 2 megohms, prices are less switch, \(2 / 6\) each ; single-pole switch \(3 / 9\) each.

\section*{TUNING CONDENSERS}

2-gang . 0005 mfd with long spindie ....................................... 5
Ditto, but fitted with trimmers .....................................

Single-gang .0005 standard size with medium-length spindle.... \(\quad 2\)

\begin{abstract}


Polished walnut radio cabinet size \(20 \times 12 \times 7 \frac{1}{1} \mathrm{in}\). complete with L., M. and S. dial, size \(7 \times 6\) in. and backplate with magic eye cutout, also with drilled chassis and hardboard back. You will find it quite a simple matter to complete this into a very handsome receiver of the \(£_{5} 5\) class. Limited quantity, price \(32 / 6\), plus \(2 / 6\) carriage for the 5 items.
\end{abstract}

\section*{FOR YOUR LABORATORY}

You many times have felt the need of a device which would enable you to put resistance or capacity or a combination of these two quickly into a circuit. We have a small quantity of resistances and capacity boxes which, by the simple manipulation of plugs, will enable you to do this. With these boxes you can put in 1 ohm, 2 ohms, 3 ohms, 4 ohms, and so on, in steps of 1 ohm, right up to 6,000 ohms. In a similar way capacity can be In a similar way capacity can be put into circuir simple for you to thus making it simple for you to find optimum working conditions. These boxes made for Government Laboratories, are available while they last at \(19 / 6\) each plus \(1 / 6\) post and packing. Don't delay-order by return.

\section*{TWO SUPER SPEAKERS}

The first is a beautifully made 10 in . P.M. speaker, a real precision product made by a very famous product made by a very famous
speaker firm. It is undoubredly a speaker firm. It is undoubtedly a l0in. which reproduces with all the quality of a 12 in . It has three special featutes (1) a solid diecast frame (2) a special speech coil suspension which gives wider frequency response (3) dustproof cone assembly. Speech coil is normal 2.3 ohims. Price is \(16 / 6\) plus \(2 / 6\) post and insurance. The second is the 8 in . P.M. speaker made by the same firm whose name incidently we are not allowed to mention but you will recognise it immediately. This again has normal \(2 / 3 \mathrm{ohm}\) coil rice is \(11 / 6\) plus \(1 / 9\) post and insurance.

BREAK-DOWN UNIT
At present day prices the spares in this unit would cost at least \(£ 5\). Hese is a list of the main contents 3 two-metre coils;
3 tuning condensers, split-stator type;
4 two-watt carbon resistors, usefu values;
1 tapped 20 watt resistor, vitreous covered;
6 paper condensers, 05 mf . 1,000 v. working ;

3 paper condensers, 1 mf . 1,000 v. working ;

2 H.F. chokes ;
4 paper condensers, 1 mf .450 v . working ;
2 paper condensers, 15 mf .
5 bakelite moulded mica condensers . 001 ;
1 paper condenser, . 01 mf. 3,000 v. working ;

24 rubber grommets, assorted sizes 6 resistors 1 watt, all useful values ; 6 resistors \(\frac{1}{2}\) wath, all useful values;
40 resistors \(\frac{1}{\frac{1}{4}}\) watt, all useful values;
40 silver mica condensers assorted values, including: 10, 15, 20 \(40,50,100,150,300\), and 500 pf . types:
4 English octal valve holders;
2 English 5-pin valve holders;
2 English 5-pin valve holders;
1 E.F. 50 type valve holder;
3 diode valve holders;
1 louvred casing, size \(12 \times 7 \times 4 \mathrm{in}\).
1 heavy metal chassis size \(12 \times 7 \times 2\) 8 in.;
8 condenser clips assorted sizes. Also an assortment of nuts, bolt P.K., self-threading screws, tag boards, chassis mounting tag con nectors, screened grid caps, plain grid caps, levers rollers, connectling rods, outputsockets, etc., etc. ALL THIS COLLECTION OF PARTS FOR 6/6 only, plus \(1 / 9\) postage and packing.

\section*{THIS MONTH'S SNIP}

RADIOGRAM UNIT BY A VERY FAMOUS MAKER. COMprising centre drive induction motor with speed regulator, auto stop and magnetic pick up, all mounted on a full size unit plate. Brand new in manufacturer's cartons. The motors are 78 R.P.M. type, but we understand that as they are governor controlied, their speed can be reduced to 33 R.P.M. or they can be made into dual speed motor by quite a simple modification. We have no precise details of this modification at the moment, but if these come to hand we will gladly supply cation at the moment, but if these
free of extra charge to purchasers.
Only a limited supply of these radiogram units are available at this Only a limited supply of these radiogram units are avainable at this
month's Snip price of \(£ 55 \mathrm{~s}\). ea. (this is almost half proper price) so month's Snip price of \(£ 55 \mathrm{~s}\). ea. (this is almost halif proper price) so
order by return. If not calling enclose \(3 / 6\) extra for packing and insurance.

15 valves. Large 12in. black
15 valves. Large 12in. black

\section*{FHRST AGAIN!}

12" TELEVISIDN AT THE LOWEST PRICE

First Time Offered

\section*{T.V.CHASSIS - LONDON or BIRMINGHAM} NEW AND UNUSED. This is not a kit but a factory built job. Fully assembled and tested, made by well-known manufacturer. Sound and vision RF. unit uses latest type miniature valves, 6 of 6AM6 and 2 of 6AL5. Permanent magnet focussing.
Complete with all valves. Any type of 12 in . c.r. tube can be used.

High quality sound with Elac 8in. p.m. speaker. Closed field.
Hard valve time bases on frame and line, uses 27 C 5 and 162 BT .
Line fly back E.H.T. 7-8 Kv., non lethal using EY51.

\section*{}

Complete with 15 valves, but less cathode ray tube.
Carriage and insurance \(37 / 6\) extra.
Power supplies \(200-250\) volts 50 c.p.s. Auto wound mains transformer, with two 6 volt, a 2 volt and 5 volt windings for valve heaters. 53 KU rect. Dimensions: \(15 \frac{1}{2} \mathrm{in}\). wide, \(15 \frac{1}{\mathrm{~h}} \mathrm{in}\). deep, 10 in . to top of tube support. Make no mistake, this is no Ex-Government lash up, but a new television receiver, complete in every detail, and when fitted with a cathode ray tube it is ready to function. Mount it in a cabinet of your own choosing, in a wall or any location that meets your requirements. Complete working and circuit diagrams, for future reference supplied with each receiver. Write for further details.
Better still call and view. "SEEING IS BELIEVING."
 LASKY'S RADIO
370 HARROW ROAD, PADIDINGTON, LONDON, W. 9 Telephone: CUNningham 1979
Write for a copy of our current Bulletin giving full details of our stocks of new manufacturars surplus components, valves, transformers, etc., also Ex-Government equipment. Enclose a \(2 \frac{1}{2} d\). stamp, with your name and address, for a copy by return.

\section*{AT LAST IT'S HERE!}

A Portable Amalyser at arice You Can Afford

\section*{THE L.S.L. VALVE AND CIRCUIT ANALYSER Mooern \(S_{\text {enncime }}\) Менноo}


Six important features :-
1. The resistance of the meter is at the rate of \(2,000 \mathrm{ohms}\) per volt.
2. Used as an external meter it has a total of 30 ranges of voltage, current and resistance readings.
3. It is the only universal meter on the market which employs two moving coil meters on the one panel, thus giving a simultaneous indication of current and voltage reading.
4. The meter is extremely useful in locating difficult and intermittent faults by leaving it in circuit in the stage suspected on a "soak " test, the meter automatically indicating incorrect readings when they occur.
5. It is the only meter on the market possessing an internal switch whereby "zero end" can be carrieo out on the three resistance ranges by a flick of the switch and without the bother of holding two external wires together.
6. There is a current reversing switch included which is particularly valuable in measuring grid bias.

LASKY'S PRICE 110196. With leads and adaptors. Carriage and insurance \(10 /\) - extra.

\section*{AND NOW A TELEVISION/SOUND VISION UNIT. NEEDS NO CONVERSION}

Model A-London. B-Birmingham. By well-known British manufacturer. Uses 10 of the latest type miniature valves. Specification and line up: Vision. 3 R.F. stages with 6AM6's. Det. and int. suppr. 6AL5. Syn. Sep. 6AM6. Video output 6AM6. Sound: 3 R.F. stages with 6AM6's. Det. and int. suppr. 6AL5. Sound output 6AM6. Size of unit : \(10 \mathrm{in} . \times 4 \mathrm{if} . \times 1 \mathrm{in}\). Circuits and full data supplied with each unit. Remember these are brand new manufa turers surplus and not Ex-Government. COMPLETE WITH VALVES. | LASKY'S PR CE f6/19/6. | Carriage 2/6 extra.


MAINS TRANSFORMERS.
Primary 200-250 volts 50 c.p.s.
MBA/3. 350-0-350v \(80 \mathrm{~m} / \mathrm{a}\), 6.3 v 3 a . 5 v 2 a .

Both filament windings tapped at 4 v .
PRICE 18/- Post Free.


SBT/1. 300-0-300v 80m/a.,
\(6.3 \mathrm{v} 3 \mathrm{a} ., 5 \mathrm{v} 2 \mathrm{a}\).
PRICE 15/- Post Free.


ANTENNA RELAY UNITS
TYPE CBY/29125
Contains \(0-5 \mathrm{~m} / \mathrm{a} \mathrm{m} / \mathrm{c}\) meter, relay insulators 0.75 amp heating element and switch. Also 50 pf. 5 Kv vacuum condenser. In black crackle metal case as illustrated.

LASKY'S PRICE:
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\hline 2 mfd . & 350 v.w. & 1/- & 25 mfd . & 25 v.w. & 1/6 \\
\hline 4 & 350 & 1/3 & 50 & 12 & 1/- \\
\hline 4 & 500 & 2/- & 64 & 450 & 5/11 \\
\hline 8 & 350 & 2/- & \(8 \times 8\) & 150 & 2/6 \\
\hline 8 & 500 & 2/6 & \(8 \times 8\) & 500 & 4/6 \\
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\hline 16 & 500 & 3/- & \(8 \times 16\) & 500 & 4/II \\
\hline 20 & 12 & 1/- & \(16 \times 32\) & 350 & 4/II \\
\hline & & & \(32 \times 32\) & 350 & 4/II \\
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CWPLIFIER ACIII: We are pleased to announce that we are now using a Varley Mains Transformer, specially wound for us, and a Varley D.P. 47 Hi -Fi Output Transformer in this popular kit. Specification: 7 valves (incl. rectifier). Push-pull Output up to 10 watts \((2 \times 6 \mathrm{~V} 6 \mathrm{G})\) to \(3 \Omega\) or \(15 \Omega\) Speaker (state which). Supply available for suner unit, exc., 280 v. 30 mA approx. High and Low Gain Inputs. Separate Bass and Treble controls. Negative Feedback. Complete kit, with eircuits and inseructions (incl. circuit of suitable quality local-station runer). E7/17/6.
"OSMOR" GLASS DIAL ASSEMBLY: 3-colour, 3-wave, 7in. \(x\) 7in., with drive mechanism. 25/-
"OSMOR" "Q" COILPACKS: A selective high-Q coilpack of miniature size ( \(1 \frac{3}{6} \mathrm{in}\). \(\times 3 \frac{1}{2} \mathrm{in}, \times 2 \mathrm{in}\).) with easy one-hole fixing. Type HO, Long, Medium and Short, \(33 /\) - plus \(7 / 4\) P.T. Type LM, Long. Medium and Gram., 28/- plus 6/4 P T Type TB, Medium Long and Shipping, \(35 /\)-plus \(7 / 10\) P.T. Type EX, Medium, Short and Short, \(35 /\) plus \(7 / 10\) P.T. Type B, Long, Medium and \(\$\) hort Battery with frame aerial, \(37 / 6\) plus \(8 / 4\) P.T. Type T.R.F. Long, Medium for T.R.F. Receivers, 30/-plus \(6 / 8\) P.T.
"OSMOR" H.F. STAGES: Designed for use with above coilpacks. For HO, TB and EX Coilpacks, \(15 /\) - plus \(3 / 4\) P.T. For LM Coilpack, 13/-plus 2/10 P.T.
I.F. TRANSFORMERS: 465 k.c., seandard size, capaciey-euned Made for us by a leading manufacturer. Per pair, 12/6.
T.R.F. COILS: M. \& L. wave, with reaction winding. Per pair, 6/6.
WEARITE "P" COILS: Full range in stock. Each 3/.
MAINS TRANSFORMERS: Pri: \(0-200-220-240 \mathrm{v}\). Sec: \(350-0-\) 350 v. \(80 \mathrm{~mA} .0-4-6,3\) v. 4A. 0-4-5 v. 2 A. Electrostatic screen. Tag connections. Universal mounting. A truly all-purpose transformer. Each \(16 / 6\)
MAINS TRANSFORMERS: Pri: 0-200-220-240 v.Sec. 250-0-250 v. \(80 \mathrm{~mA}, 0-4-6.3\) v. 4 A . \(0-4-5 \mathrm{v}, 2 \mathrm{~A}\). Electrostatic screen. Tag connection.. Universal mounting. Each \(16 / 6\).
BALANCED ARMATURE HEADPHONES: Type DLR2. With wide, comforrable double metal headband, lead and plug. The most sensitive low resistance phones made. Boxed. Per pair T/-iIMMERS: - Miniature ceramic air-spaced 3-30pF 5d, each, \(4 / 6\) per doz. Phi.ips Concentric type: \(3-30 \mathrm{pF}, 6 \mathrm{~d}\). each, \(5 / 6\) per doz., 3 per gross. Ceramic air-spaced 75pF \(1 /=\) each. Compression Ceramic type; 50 pF 6d. each, \(100+100 \mathrm{pF}\), 9 d . each, \(250 \mathrm{pF}, 1 /\) - each, 500 pF, \(1 / 3\) each, 1,000 pF, 9 d . each.
ALADDIN COIL-FORMERS: Type F804, with iron-dust cores, 5d. each, \(4 / 6\) per doz., \(48 /\)-per gross.
THERMALLY OPERATED MICRO-SWITCHES: Normally "on." Break at approx. \(100^{\circ} \mathrm{C}\). (adjustable). Control \(10 \mathrm{~A}, 250 \mathrm{v}\). Each 3/6.
ELECTROLYTICS: By leading makers. All new and guaranteed, \(1 / 350,2 / 6 ; 2 / 200,1 /-; 2 / 350,1 / 6 ; 4 / 200,1 / 3 ; 4 / 450,2 / 6 ; 8 / 350,2 / 6 ;\) \(8 / 450,3 /-; 8 / 500,3 / 3 ; 16 / 350,2 / 6 ; 16 / 450,3 / 6 ; 16 / 500\) 4/-; 4/9; 5/-; \(32 / 350,3 /-; 32 / 450(\mathrm{~T} / \mathrm{V}), 6 /-; 8-8 / 450,3 / 6 ; 9-8 / 500,4 / 8 ; 8 \cdot 16 / 450\), 42/6 and \(5 /-; 8-16 / 500,6 /=; 16-16 / 350,4 /-; 16 / 16 / 450,6 /-; 16-32 / 350,5 /-\) \(\begin{array}{ll}4 / 6 \\ 32-32 / 450 ; & 6 / 6 ; 60-1 /-160 / 350,6 / 350,4 /-; 16 / 16 / 450,6 /-; 16-32 / 550,5 /-; \\ 50 / 50,1 / 9 ; 20 / 12,2 / 6 ; 25 / 25,1 / 9 ;\end{array}\) \(32-32 / 450,6 / 6 ; 60-100 / 350,6 /-;\)
\(50 / 12,1 / 9 ; 50 / 25.2-; 50 / 50,2 / 3\).
OIL TEMPERATURE INDICATORS: A sensitive movingcoil movement, F.S.D. of which is approx. \(120 \mu \mathrm{~A}\). Centre-tapped moving coil. Calibrared \(0-120\). Ideal as indicator in bridges, signal strength meter, valve-voltmeter movement, or can be modified as sensitive relay. \(3 /\)-each, \(30 /\) - per doz.
EHT CONDE NSERS:
EHT CONDENSERS: 0015 mid. 8 kV wkg. Ideal for EHT from line flyback. \(2 / 6\) each. 01 mfd .5 kV wkg. \(1 / 6\) each.
RESISTORS: \(\ddagger \mathrm{w}\). , \(3 \mathrm{~d} . ; \frac{1}{2} \mathrm{w} .4 \mathrm{~d} . ; 1 \mathrm{w} ., 6 \mathrm{~d} . ; 2 \mathrm{w} .9 \mathrm{~d} . ;\) Wide range of values in stock.
WIREWOUND RESISTORS: 6 wart, \(2,24,4,5,14,14,31.5\) ohm, I/- each. \(12 / 15\) watt, 100 , \(1 \mathrm{k}, 7.5 \mathrm{k} ., 9{ }^{\prime}\) k., il k., 12.5 k., 30 k . ohm, \(1 / 6\) each. 45 wate, \(80 \mathrm{ohm}, 18 \mathrm{k}, 2 / 3\) each. 100 watt, 800 S , 1 k ., \(10 \mathrm{k}, 20 \mathrm{k}, 3 /\) - each. 200 watt, \(430 \mathrm{ohms}, 4 /\) - each.
HEADPHONES, TYPE DLRI: Low Resistance, with headband, lead and plug. New and unused. Special offer at \(3 / 9\) per pair, or 42/- per dozen pairs.
CONDENSERS: Ceramic, silver mica, mourded mica, in following values (al! pF's). 2, 4, 10, 15,20,22,25, 47, 50, 100, 170, 200, 220. \(250,300,330,500,1,000\), 6d, each, 5/-per doz.
TOGGLE SWITCHES: D.P. on-off, laminared bakelite, \(1 / 6\). D.P.D.T., black bakelite, 2/6. D.P. on-off, black bakelite, ex-equipment, \(1 /-\)
NEONS: 85 v . striking, no eesistor in cap. With holder, \(2 /-\) each, 21/- per doz.
RUBY LAMPHOLDERS: To take M.E.S. bulb, \(1 /\)-each.
"CLEAREX" TELEVISION LENSES: For 9in. tube, 50/, with filter \(55 / \%\). For 12 in . Tube, \(70 \%\), with filter, \(75 /\). TELEVISION FILTERS for 9 in. tube \(18 /\). For 12 in . tube \(21 /\). STABILIVOLTS: Type NSI. Voltage stabiliser and divider. Operating voltages \(280 \mathrm{v} ., 210 \mathrm{v}, 140 \mathrm{v}, 70 \mathrm{v}\). Max. electrode current 80 mA . Each \(10 \%\).
MICROPHONE TRANSFORMERS: Individually boxed. Each 2/6.
METER OFFER: \(2 \frac{1}{\mathrm{i}} \mathrm{in}\). Meters, FSD 50 AA D.C. Tropicalised. Erand new and unused. Few only ar 30/- each.
JUNEERO TOOL: Bends, shears, punches and threads. For all the litele jobs that waste your time. As reviewed in this journal. Each \(18 / 6\).
XACTO SLIDEGAUGE AND PROTRACTOR. For use with above Juneero tool \(7 / 6\).
EXTENSION SPEAKERS: Fited Sin. Plessey P.M. Speaker in bakelite cabinet, with perforated back. Also ideal for intercom. bakelite ca
Each 19/6.
AERIAL RODS: Heavily copper-plated steet, one foot long. Fit into one another to make any length aerial. Per doz. 2 PLUGS AND SOCKETS: Pye angle, \(1 /-\); Pye straight, \(1 / 3\); B. \& L. 5 pin, \(1 / 6\); B. \& L. 7 pin, \(1 / 9\); B. \& L. 10 pin, \(2 /\)-; Jones 6 way, \(1 / 6\); Jones 8 way, \(1 / 9\); Jones 10 way, \(2 /\)-; Jones 12 way, \(2 / 6\); E.H.T. Single, \(1 /\). The above prices include plug and socket in all cases. Octal Plug and Socket (Valveholder), black bakelite, I/9. Jack plug with Igranic socket, \(2 / 6\). Pye T-pieces, 6d. Pye connectors, \(6 d\).
OVERLOAD CUTOUTS: Adjustable from 50 to 100 mA . Ex-equipment, 4/-each.
MUIRHEAD DIALS: As used on G-units and R.I224A. Each \(6 / 6\).
VIEWMASTER " TELEVISION: Instructions, 5/-. Components supplied separately, or in kits as follows: Whiteley. 125/-. T.C.C. London, 135/-. Midland, 140/-. Westinghouse, \(62 / 6\); Morganite (inci. V/C's) London, 35/9, Midland, 35/-, Plessey, \(112 / 6\), Colvern, 19/3. Wearite Coils, London, 22/-. Midland, \(30 /\)
"ELECTRONIC" TELEVISION: London Handbook, 2/6. Wiring diagram, 2/6. Midland Handbook and Wiring Diagram, 4/6. Chassis, with valvehoiders, coil formers, etc., mounted: Vision, 22/6; Sound, 22/6; T/Base, 17/6; P/Unit, 25/-; Gantries, etc., \(9 /-\) Designer approved components: Coils and chokes, London, 151 -, Midland, 17/6. (Please specify London or Midland.) Focus, 30/: Scanning, 25/6. L/O Transformers, 25/6. Mains Transformer, 90/. 4 kV Transformer, \(67 / 6\). 5 kV Transformer, \(72 / 6\).
L.F. CHOKES: 350 ohm \(40 \mathrm{~mA}, 4 / \mathrm{F} ; 5 \mathrm{H} 80 \mathrm{~mA}, 4 / 6 ; 10 \mathrm{H} 80 \mathrm{~mA}\). 7/6; 3H \(200 \mathrm{~mA}, 4 / 6 ; 20 \mathrm{H} 80 \mathrm{~mA}, 8 / 6\).
P.M. SPEAKERS: New and unused, by leading manufacturers. 5 in . \(11 / 6 ; 6 \frac{1}{2} \mathrm{in}\)., \(13 / 6\); 8 in ., \(15 \mathrm{in} ., 10 \mathrm{in}\)., \(19 / 6\). All 3.
THROAT MICROPHONES: Magnetic, sultable electronic musical instruments. \(1 / 9\) per pair,
MAINS DROPPERS: \(2 \mathrm{amp}, 3 / 3 ; .3 \mathrm{mp}, 4 / 6\); Vitreous enamelled or cement-coated.
VARIABLE CONDENSERS: 50 pF S.G., I/-; 100 pF S.G., 2/-; 160 pF S.G., 2/3; 200 pF S.G., 2/6; 500 pF Two Gang, 6/6; 500 pF Three Gang, 7/6. Split Stator \(100+100 \mathrm{pF} 2 /-\); Split Stator 250 pF with worm gear and switch, 2/6.
MAT RESISTORS: 100 ohm 250 watt, \(1 / 6 ; 300\) ohm 250 watt, I/9.
VIBRATORS: 6 volt non-synchronous ( 4 pin UX), 5/-; 6 v . synchronous ( 6 pin UX), \(7 / 6\).
SELENIUM RECTIFIERS:
SELENIUM RECTIFIERS: 140 v. 600 mA (bridge) or 280 v . 300 mA half-wave, \(10 / \mathrm{F}\); 24 V .1 .2 A (bridge), \(12 / 6 ; 12 \mathrm{v} .4 \mathrm{~A}\) (bridge), \(15 \%\)
METERS: 2 din \(0-150\) v. A.C. (M.1.) American, 6/6; 2 in .50 mA D.C. (M.C.) square flush, \(61-; 2 \mathrm{in} .200 \mathrm{~mA} \mathrm{D.C}\). (M.C.) square flush, \(6 /-2 \mathrm{in} .500 \mu\) A D.C. (M.C.) \(7 / 6\).
WIRE-WOUND POTENTIOMETERS: 50 2, 200 \(\Omega, 250 \Omega 2\), \(1 \mathrm{k} \Omega .5 \mathrm{k} \Omega, 8 \mathrm{k} \Omega, 10 \mathrm{k} \Omega, 20 \mathrm{k} \Omega\), all 5 w . bakelite cased, \(2 / 6\) each. \(100 \Omega\), \(500 \Omega\). 5 K , miniature I w. preset, \(1 / 6\) each. \(50 \Omega 25 \mathrm{w} .4 / 6\) each. 17 ks 100 w . toroidal ceramic, \(8 / 6\) each. \(20 \mathrm{k} \Omega 10 \mathrm{w}\). De Jur Amsco Precision, \(4 / 6\) each.
CARBON POTENTIOMETERS: \(500 \Omega, 3 \mathrm{k} \Omega, 10 \mathrm{k} \Omega, 25 \mathrm{k} \Omega\), \(50 \mathrm{k} \Omega, 100 \mathrm{k} \Omega, 250 \mathrm{k} \Omega, 500 \mathrm{k} \Omega, 1 \mathrm{~m} \Omega, 2 \mathrm{~m} \Omega, 3 \mathrm{~m} \Omega, 1 / 6\) each.
ANTI-VIBRATION MOUNTINGS: To take 56 lb . weight, i/-
MINIATURE 24 v. D.C. MOTORS: \(2 \neq \mathrm{in}\), \(\times 1 \frac{1}{2} \mathrm{in} . \times 1 \frac{1}{2} \mathrm{in}\)., each MICA T/X CONDENSERS: \(50 \mathrm{mpF}, 1.5 \mathrm{~A}\) at \(4-7.5 \mathrm{Mc} / \mathrm{s}\). Test MICA T/X CONDENSERS: 500 pF ,
\(5,000 \mathrm{v}\) A.C. for ten minutes. 6 for 51 . 5,000 V. A.C. for ten minutes. 6 for \(5 /\) IRPOSE: \(803,805,10 /\); 8012, CV296, 7/6; VUIII, VUI20A, CV66 6/6; VUI33, CV54, 5/: 7193, 3/6.

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Primary 200-250 v. P. \& P. on each 1/6 extra.
\(250-0.250100 \mathrm{~mA} ., 6 \mathrm{v} .3 \mathrm{amp} .\), 5 v. 2 amp., Upright mounting. 17/6.
250-0-250, \(60 \mathrm{~mA} . .6 \mathrm{~V} .4 \mathrm{amp}\). (to be used on common heater chain with \(6 \times 5\) rectifier), \(13 / 6\). \(280-0-280,80 \mathrm{~mA} ., 6 \mathrm{v} .3 \mathrm{amp}\). 4 v. 2 amp., drop-through, \(14 /\)-. 4 V .2 amp., drop-through,
Same as above, but \(350-0.350\) Same
\(14 /\)..
Drop thro \(350-0-350\) v. 70 mA ., 6 v. \(2.5 \mathrm{amp} ., 5 \mathrm{v} .2 \mathrm{amp} ., 13 / 6^{\prime}\). Semi-shrouded, drop-thro' \(280-\) \(0-280.6 \mathrm{v} .3 \mathrm{amp} ., 5 \mathrm{v} .2 \mathrm{amp} ., 15 /-\) Semi-shrouded, drop-thro or upright mounting 280-0-280 80 \(\mathrm{mA},. 4 \mathrm{v} .6 \mathrm{amp},. 4 \mathrm{v} .2 \mathrm{amp}\)., 12/6.
Auto-wound H.T. 280 volts at \(360 \mathrm{~mA} ., 4 \mathrm{v} .3 \mathrm{amp} ., 2 \mathrm{v} .3 \mathrm{amp}\)., or 6 v .3 amp . Separate 4 v .3 amp . rectifier winding (upright or drop-thro'), 10/6.
Heater Transformers Pri. 200250 v., 6 v. \(1 \frac{1}{7}\) amp., \(6 /-; 2,4\) or


\section*{ELECTROLYTIC CONDENSERS}

50 mid .50 work. \(1 / 9\). \(16-24\) mid. 350 work. \(3 / 3\). 100 mfd .12 v . work. \(1 / 3\). \(16-16 \mathrm{mfd} .450\) work. \(4 /\)-. 50 mfd .12 v . wark. \(1 /\)-. 25 mid .25 v . work. \(1 / 2\). \(16 \times 8 \mathrm{mfd} .450\) work. \(3 / 9\). 8 mfd .450 v . work. \(1 / 11\). 250 mfd .12 v . work. \(1 / 3\). \(8 \mathrm{mfd} .500 \mathrm{v} . \quad 2 / 9\).
\(16 \mathrm{mid} .500 \mathrm{v} .3 / 6\).
\(8 \times 8 \mathrm{mfd} .450\) work. \(3 / 6\).
32 mfd .350 work. \(2 /\)-.
\(32+32\) M.F.D. small tube tag ends 200 v . work. at \(2 /\) -

\section*{P.M. SPEAKERS}
\begin{tabular}{|c|c|c|}
\hline Size & & less trans \\
\hline \(3 \frac{1}{2} \mathrm{in}\). & - & 9/- \\
\hline Sin. & 13/6 & - \\
\hline \(6 \frac{1}{2} \mathrm{in}\). & 13/6 & 10/- \\
\hline 8 in . & 15/6 & 13/6 \\
\hline 10 in . & \(19 / 6\) & 17/6 \\
\hline
\end{tabular}

\section*{ENERGISED SPEAKERS}

Bin. 2,000 ohm field with O.P. trans. 5,000 ohm imp., \(15 / 6\). \(5 \mathrm{in}, 1,000 \mathrm{ohm}\) field with O.P. trans., \(13 / 6\).

Post and packing 1/..

MIDGET BAKELITE CABINET, \(7 \mathrm{in} . \times 5 \frac{1}{2} \mathrm{in} . \times 5 \mathrm{in} . \mathrm{c} / \mathrm{w}, \mathrm{S}\)-valve \(\mathrm{S} / \mathrm{H}\). chassis med./long wave seale and back (Takes std, twin gang condenser and \(3 \frac{1}{2}\) in speaker), 15/-.
EXTENSION SPEAKER, 4 in . wide \(\times 4 \frac{1}{2} \mathrm{in}\). \(h \mathrm{igh} \times 3 \mathrm{in}\). deep at base tapering to 2 tin. \(\mathrm{c} / \mathrm{w}\). 3 tin. speaker. \(15 /\)-.
Colours of both the above: Cream, walnut, black and maroon. State second choice when ordering.
LINE AND FRAME COIL ASSEMBLY. Frame coils wound but not fitted (full instructions supplied). High impedance frame; low impedance line matching 5-1. 7/6.
\(6 \frac{1}{2}\) in. SPEAKER (P.M.) specially made for Television with closed field, complete with O.P. transformer. \(11 / 6\).
CONSTRUCTOR'S PARCEL
CONSTRUCTOR'S PARCEL Comprising 5 -valve superhet chassis with transformer cut-out, size 13 in \(\times\)
6 in. \(\times 2\) iñ., with L.M. and S . scale, size 7 in. \(\times 5\) in. Back plate two supporting brackets, drive drum, pointer, two-speed spindle, twin gang condenser. Mains transformer \(250-0-250\) v. \(60 \mathrm{~mA}, 6 \mathrm{v} .4 \mathrm{amp}\). Pri. \(200-250,6 \frac{1}{3} \mathrm{in}\). speaker and \(6 \times 5\) rect.
\(28 /=\) Plus \(2 /-\) post and packing.
6.STATION SWITCHED SUPERHET COIL UNIT, by famous manu facturer. Ideal for Car Radio or radio set. Range coverage Pos. \(1,200-300 \mathrm{~m}\). \(2,250-360 \mathrm{~m} . ; 3,250-360 \mathrm{~m}, ; 4,320-460 \mathrm{~m}\). ; \(5,400-550 \mathrm{~m} . ; 6,1,100-1,850 \mathrm{~m}\). no oscillator required for lining up, complete with Circuit and medium and long wave frame aerial. 21/-, post and packing, 1/-.
PRE-ALIGNED MIDGET 465 Kc . Q. 120 made for the above Coil Unit. \(8 / 6\) per pair, post 6 d
CHASSIS TO FIT COIL UNITS AND I.F.S, size \(11 \frac{1}{2}\) in. \(\times 5 \frac{1}{6} \mathrm{in} . \times 1 \frac{7}{6} \mathrm{in}\). 2/6
MAINS TRANS. TO FIT ABOVE CHASSIS. Pri. \(200 / 250\) volt. Sec \(250-0-250 \mathrm{v} .60 \mathrm{~mA} ., 6 \mathrm{v} .4 \mathrm{amp} ., 13 / 6\), post and packing, 1/-. CONSTR UCTOR'S PARCEL, comprising :-
Chassis 10 tin. \(\times 5 \frac{1}{2} \mathrm{in} . \times 2 \mathrm{in}\), with speaker and valve holder cut-outs, \(R\), and \(A\). \(6 \frac{1}{2}\) in. P.M. with transformer, twin gang with feet, pair medium and long wave iron cored T.R.F. coils. Four International Octal valve holders, wave-change switch and Erie 20 k pot with switch, \(25 /\)-, plus \(1 / 6\) post and packing
CONSTRUCTOR'S PARCEL, comprising chassis \(10+\mathrm{tin} . \times 5 \frac{1}{2} \mathrm{in} . \times 2 \mathrm{in}\). with speaker and valve holder cut-outs; Rola 5 in . P.M. with O.P. trans, ; twinnang with trimmers; pair of T.R.F. coils; Four international Octal valve holders : wave change switch and Erie 20 k pot, with switch, \(19 / 6\), plus \(1 / 6\) post and packing. CONSTRUCTOR'S PARCEL, comprising Midget twin-gang with slowmotion drive ; pair midget 465 Kc . I.F.s ; frame aerial ; medium wave osc. coil and layer type H.T. and L.T. batteries \(90 \mathrm{v} .+1 \frac{1}{3} \mathrm{v} ., 21 /-\) plus \(2 /\)-postand packing. STANDARD 465 KC. I.F.s. Iron cored Q.120, 7/- per pair.
MINIATURE 465 KC. I.F.s. Type M400B, 12/6, plus 6d, post and packing. MINIATURE 465 KC. I.F.s (slightly larger than the above item), Q. 120. Per pair 10/-.
Per pair 1 ORED 465 KC . I.F.s. Q. \(130,2 \frac{3}{3} \mathrm{in} . \times 1 \frac{1}{6} \mathrm{in}\)., per pair \(6 /\) -
IRON CORED 465 KC. WHISTLE FILTER, screened, each \(2 /\) -
VALVE HOLDERS. Paxolin International octal. 4d, each. Moulded International octal, 6d, each. EF50 ceramic 7d, each. Moulded B7G slightly
soiled 6d, each. 3-way 0.3 amp ., 180 ohm per yard. Ild. per yard. 3-way
LINE CORD. \(0.2 \mathrm{amp}, 300\) ohms per yard, \(1 /\)-.
CERAMIC P.F.S. 3 each of the following: 330,220180 and \(82,2 / 6\).
VOLUME CONTROLS, by famous manufacturer. Long spindle and switch, \(\ddagger, \frac{1}{2}, 1\) and 2 meg, \(3 / 6\) each. 20,25 and \(50 \mathrm{k} ., 3 /\) - each. Post and packswitch, \(\frac{1}{d}, \frac{1}{2}, 1\)
ing 3 d . each.
ing 3d. each. CONTROLS, by famous manuíacturer. Long spindle less switch, 5 k .50 k .500 k .1 t meg., 1/9 each. Post and packing 3d. each.
5-VALVE A.C. MAINS, \(200-250\) V., 3 wave-band superhet chassis. 1,000-\(2,000,200-550\) and \(16-50\) metres. By very good manulacturer, complete with valves. \(6 \mathrm{~K} 8 \quad 6 \mathrm{K7} 6 \mathrm{G} 7 \quad 6 \mathrm{~V} 6\) and \(5 \mathrm{Z4}\). Size of chassis \(111 \mathrm{in}, \times 7 \mathrm{tin} . \times 2 \mathrm{in}\). Size of glass scale \(8 \frac{3}{3} \mathrm{in} . \times 7 \frac{1}{2} \mathrm{in}\). with pick-up sockets (takes 8 in . P.M. with trans.). TAX PAID. \(88 / 18 / 6\). Post and packing \(2 / 6\).
CONSTRUCTOR'S PARCEL, comprising, 5 -valve superhei chassis with I.F. trans... and V/H. cut-outs, \(13 \frac{3}{4} \mathrm{in} . \times 6 \mathrm{in} . \times 2 \mathrm{in}\), with L.M.S. scale 7 in . \(\times 5 \mathrm{5in}\)., back plate and two supporting brackets. Drive drum, spindle and pointer, mains trans. G.E.C. \(350-0-35070 \mathrm{mA} .6 \mathrm{v} .2 .5 \mathrm{amp} ., 5 \mathrm{v} .2 \mathrm{amp} .\), and Plessey 8 in . M. energised \(2,000 \mathrm{ohm}\) field with \(5,000 \mathrm{ohm}\) impedance, O.P. trans., 2916. Post and packing \(2 /\)-.

CONSTR UCTOR'S PARCEL, comprising: 5 -valve superhet chassis with I.F. trans., and V/H. cut-outs, \(13 \frac{3}{3} \mathrm{in} . \times 6 \mathrm{in} . \times 2 \mathrm{in}\). with L.M.S. scaie \(7 \mathrm{in} . \times 5 \mathrm{in}\)., back plate and two supporting brackets. Drive drum, spindle and pointer, back plate and two suppor
CONSTRUCTOR'S PARCEL, comprising: twin gang with ieet, pair long and medium iron cored T.R.F. coils, 4 octal V/H., 3 pole 2 -way switch Erie 20k. pot.w/s., heater trans. \(220-240 \mathrm{v}\). 1 amp., 2 yds. flat silk flex, 3 black knobs, pot. W/s in heater trans. 220 . with speaker and \(V / H\). cut-outs, metal rectifier chassis 101 in . \(\times 5 \frac{1}{4} \mathrm{in} . \times 2 \mathrm{in}\). with speaker and 20 m .
60 mA . 230 v . and \(32+32 \mathrm{mfd} .200 \mathrm{v}\)., \(19 / 6\), post paid.
GOVERNMENT SURPLUS. Removed from equipment but guaranteed Metal Rectifier, \(230 \mathrm{v} .60 \mathrm{~mA} ., 4 /\) -
Metal Rectifier, \(230 \mathrm{v} .80 \mathrm{~mA} ., 5 /-\).
Post and packing on each, 6d.
8 mfd . metal cased, 450 v . work, with elip, \(1 /\) - each.
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2 2mp. 1,000 ohms, tapped 900 ohms. 1/9.
.2 amp. 717 ohms, tapped 100 ohms. 1/6.
.3 amp. 520 ohms, tarped \(2 / 6\) each.

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3-pole 2 -way, \(1 / 2\).
4-pole 4-way, and 4-pole 3-way 1/9.
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White Rubber. 9 in. with glass, \(10 / 6\).

DUETOA LARGE PURCHASE OF MANUFACTURERS' SURPLUS WE ARE ABLE TO OFFER
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Heater transformer fully interleaved and impregnated and guaranteed. 6.3 V. \(\quad\) amp,
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CONSTRUCTOR'S PARCEL, comprising chassis with V.H. and transformer cut-out, size \(13 \frac{1}{4} i n . x 6 i n . x 2 i n\). with sca.e size 7 in.x5in. Back plate, wo supporting brackets, drive drum, pointer and spindle. Mains transformer input \(200-250\) v. H.T. 280 v. 4 v. 3 amp .4 v .3 amp . and 3 valves, 2 Octai H.F. pentodes, and I output valve, all ex-Govt. 2ti 1s. Od. Post paid.

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\hline Tannov 12 volt Loud Hallers, complete with control box, mike and speaker & ¢6 & 15 & 0 \\
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complete with resistance and
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Avo Valve Tester, \(1948 / 9\) roller panel model, complete and per fect .................................................. Avo Model 7, as new ................ 1110 Avo Model 7 with Avo leather case, as new
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operation. In new condition....
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65B, as new \(\ldots . . . . . . . . . . . . . . .\).
Taylor Valve Tester, and meter
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combined model 47 AP , as new... \(\subset 1810 \quad 0\)

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215, as new
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A.C. only, as new .................... \&19 0

Collaro Microgram Auto Changer Player, as new ....... \&21 100
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WE HAVE OTHER EQUIPMENT ARRIVING DAILY!

Labgear Electronic Fault Tracer, in new condition
Taylor Valve Tester Model 45A complete with chares. As new. 1110 Denco Communication Receiver, \(200 \mathrm{kc} / \mathrm{s}\) to \(40 \mathrm{mc} / \mathrm{s}\). In new condition. Complete and perfect
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Garrard RC65 Ul6c Universal record changer, AC/DC, mixed IOin, and 12 in . Brand-nyw..... ©19 100 Transmitters No. 12, Brand new, complete and perfect ...... 22 0 0 Transceiver No, 58, complete with power unit, new ............. \(\$ 1210\)

\section*{CLYDESDALE}

Bargains in Ex-Service Radio and Electronic Equipment


估 superhet receiver provides yet another basis for that 2 metre rig.
Transmitters prove comparatively simple to build but the receiver is a nother story, but let us expla inWe can supply you with this STRATTON built receiver which can be modified for use on " 2 ." The set in its present state can be tuned to any spot frequency between 85 and \(95 \mathrm{Mc} / \mathrm{s}\) by means of a crystal controlled oscillator stage which should not prove unduly difficult to alter, and pruning the coils is quite simple. REMEMBER ! The P40 would make a FB mobile recelver for " 144 ,"
Emoloying an I.F. of \(2.9 \mathrm{Mc} / \mathrm{s}\), the valve line includes: I/EF54 RF, I/EF54 MIXER, I/EC52-2/EF54 OSCIL: LATOR \& MULTIPLIER STAGE, 2/EF39 IF, I/E834 DET \& AVC, \(1 / 655\) Ise audio, and a 6 V 6 ourpue. The complete unit is enclosed in a die cast frame with The complete unit is enclosed in a die cast ira
a louvred cover. Dimensions \(11 \frac{1}{2} \times 5 \frac{2}{2} \times 4 \frac{1}{2} \mathrm{in}\).
CLYDESDALE'S PRICE 23.19 .6 ach POST
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\section*{POWER FOR THE P40 RX.}

THE POWER UNIT S-451-B designed for use with the P40 receiver proves to be an extremely handy little unit in lany "HAMS" shack, in that it has an LT output of 12 voles, a supply which is often needed for powering Ex-W.D. surplus equipment.

The S-451-B built by STRATTON is complete and ready for use on an A.C. malns supply of \(200 / 250\) voles \(40 / 60 \mathrm{cy}\).

The outputs are 12.5 v . at 2.5 amps and 175 v at \(60 \mathrm{M} / \mathrm{a}\). with double choke smoothing and condenser input.
The power unit is completely enclosed in a die-cast metal frame with a louvred cover. Dimensions \(11 \frac{1}{2} \times 5 \frac{1}{2} \times 4 \frac{1}{2}\) in.

CLYDESDALE'S PRICE \(39 / 6\) each
POST PAID.

\section*{FOR MARINE USE.}

The CRV 46151 is a 6 -valve superhet suitable for shipping, with a frequency coverage of \(195-9050 \mathrm{ke} / \mathrm{s} \mid\) RF., and 3 I.F. stages, plus an audio output stage. The 3rd I.F. stage provides C.W. Ose.

VALVE LINE-UP: 4/125F7, 1/12SA7, 1/12A6 and 1/991 seabiliser.
H.T. power is derived from a 28 v , dynamotor mounted sub-chassis; which has an output of \(260 \times 770 \mathrm{Ma}\). The valves are wired in series/ 260 v/lo Ma. The valves are wired in
parallel providing a \(24 \vee\) heater circuit.
Complete in metal case \(15 \frac{1}{\frac{1}{2}} \times 8 \frac{1}{\frac{1}{2}} \times 7 \frac{1}{\mathrm{t}} \mathrm{in}\). Weight 2 s .8 lb .
CLYDESOALE'S PRICE \(\$ 5.19 .6\)
CARRIAGE PAID.

THE 40 METRE RX YOU CAN TAKE WITH YOU.
The receiver unit of the No. 18 WALKIE TALKIE provides an extremely compact set for operation on " 40 " During tests on \(7 \mathrm{Mc} / \mathrm{s}\) the 18 Rx proved extremely sensitive and requires only a 2 v . accumulator and a \(90 / 120 \mathrm{v}\). H.T. supply with about \(\mathrm{I}+\mathrm{v}\), of bias. The \(18 R x\) is a 4 valve superhet with regeneration for use when C.W. reception is required, by increasing the L.F. gain control to almost MAX the receiver is in a state of oscillation. The I.F.
als is \(465 \mathrm{Kc} / \mathrm{s}\).
The valve line-up includes \(/ /\) ARP12 R.F., \(1 /\) ARP12 Mixer, \(1 / A R P 12\) I.F., and I/AR8 2cd DET \& AVC and AUDIO OUTPUT. The output jacks are provided for headphone use. Circuits and details are available as required.

> CLYDESDALE'S PRICE \(17 / 6\) ONLY 176

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\section*{HOT NEWS TIP FOR THE COMMERCIAL RADIO CONSTRUCTOR}

\section*{WALNUT WOODEN CABINETS.}

A professionally built finely finished walnut wooden cabinet with inside dimensions \(157 \times\) \(8 \frac{1}{} \times 7\) Itin. complete with a finished diala perture in the front \(5 \frac{5}{5} \times 3 \frac{1}{\mathrm{i}} \mathrm{in}\). and a speaker aperture 5fxifin. Three holes are drilled symmetrically below the dial aperture. This is the type of cabinet which will make want to build that new house set. Refer H394.

CLYDESDALE'S PRICE 81.5 .0 each
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DIAL GLASS (Graduated). H 410. A three WAVEBAND DIAL GLASS to suit the cabinet detailed above, completely graduated with station names and wavelength. Dimensions \(6 \times 4 \mathrm{in}\).

CLYDESDALE'S \(1 / 3\) each POST
PRICE ONLY PAID

CONTROL KNOBS.
IVORY CONTROL KNOBS Itin. diam., in. deep with serrated edges suitable to fit tin. spindle with flat side. Knob complete with spring ellp. Refer H403
CLYDESDALE'S
PRICE ONLY 9d. each POST PRICE ONLY

POST
PAID.
H404, as above but finished in BROWN
CLYDESDALE'S 8d each POST
PRICE ONLY 8U. PAID.

\section*{LOUDSPEAKERS.}

H396 BRAND NEW a inch PM MOVING COIL LOUDSPEAKERS complete with matching transformer. These speakers are standard replacement types and ready for installation, in original makers cartons.
CLYDESDALE'S \(17 / 6\) each POST PRICE ONLY \(17 / 0\) PAID.

\section*{A "MUST" FOR YOUR WORKBENSH.}

An inexpensive 1 H.P. motor-modified from the TYPE 29 MOTOR GENERATOR for use on \(\mathbf{2 0 0 / 2 5 0}\) Volts A.C. mains. This motor makes an ideal buff, or lighe tool grinder-or a static drilling machine, a drill chuck can be easily fitted to the end of the lin. spindle after the existing fan has been removed. Motor size Ilin. \(\times 5 \frac{1}{2}\) in.

CLYOESDALE'S PRICE ONLY \(25 /=\) each. POST PAID.

\section*{CONDENSERS.}

H399 TWO GANG variable tuning condenser less trimmers. 00035 Mid . nom, with tin, spindle \(\frac{1}{2} \mathrm{in}\). long. Dimensions \(2 \frac{i}{4} \times 2 \mathrm{f}\) in.
CLYOESOALE'S \(\mathbf{7 / 6}\) each POST
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MEDIUM and LONG. WAVE PADDERS. is0 and 4CO PI. respectively. These two trimmers are mounted in a single ceramic block fitted with a small bracket for mounting purposes.
CLYDESDALE'S 1/- each POST PRICE ONLY

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H 4008 Mfd. 450 v . Wkg. Ali-can Electrolytic condenser \(2 \frac{1}{} \times \mathrm{lin}\). diam.
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\(2 / 9\) each
POST
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ips for the above
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H \(40 \mathrm{I} 16 \times 24 \mathrm{Mid} .450 \mathrm{v}\). Wkg. Ali-can Electrolytic condenser \(2 \frac{1}{2} \times 1\) in. diam.
CLYOESDALE'S 4/6 each POST
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\(4 / 0\)
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H401A Mounting clips for the above condenser.
CLYOESDALE'S 3 d , each or \(2 / 6\) doz.
POST PAID.
 packing cases. Incorporating 15 valves sype EF50, 2 of SP61, EF36, EBC 33,3 of EB34. Complete \(45 \mathrm{me} / \mathrm{s}\). I.F. Serip, motor dial and drive pots, etc., etc. \(£ 6\) only, plus \(10 /-\) packing and carriage. Whilst they last EX-R.A.F, INDICATOR UNIT TYPE 62. Containing VCR-97 CRT with mu-metal screen: Xtal Unit and valves \(16 /\) VR65 (SP61) 2/VR54 (EB34) 2/VR92 (EASO), etc.. etc., two deck chassis in meta case. \(18 \times 18 \frac{1}{2} \times 11 \frac{1}{\mathrm{i} n}\). New condicion. \(67 / 6\) each. Pius \(7 / 6\) packing and Carriage.
INDICATOR UNIT TYPE 198. Containing VCR 138A \(3 \frac{1}{2}\) in. tube 3 VR55, I VR54, I VR92, 2 high-speed relays, volume controls and 10 res, and condensers. Absolutely brand new. Carriage paid, 2. BAKELITE RECEIVER CABINETS. An exeremely advaneageous purchase, enables us to offer the following :-Ateractive brown bakelite cabinet, size 15 in , \(\times 8 \frac{\mathrm{z}}{\mathrm{i}} \mathrm{in}\). high \(\times 7 \frac{1}{2} \mathrm{in}\). deep, complete with chassis drilled for standard five-valve superhet, back, 3 -wave glass dial and back plate. Chassis and cabinet are designed for \(6 \frac{1}{2} \mathrm{in}\). speaker, and all standard components. Price complete is 25/-only. Limited quantity. WAYE-FORM GENERATOR TYPE 34, Ex. A.M. Including 6 SP6!, 4 EF36, 2 EB34 and one CVII6. Also relays, transformers pots., condensers and resistors. The whole contained in metal box size \(\left\|\frac{1}{3} \times\right\| \times 8 \mathrm{in}\). In clean condition, an absolute bargain at \(25 /-\) plus \(3 / 6\) packing and carriage.
A.M. RECEIVER UNIT, TYPE 161. Comprising RL37, 2 EF54 and EC52. Coils, relay and many condens srs and resistors. The whole in metal box size \(8 \frac{1}{2} \mathrm{in} . \times 6 \frac{1}{3} \mathrm{in}\). \(\times 3 \frac{3}{2} \mathrm{in}\). New, a bargain at only \(15 /\) Carriage Paid.
SPECIAL PURCHASE. Brand new H"4/200 E.H.T. pencil rectifiers, \(2,400 \mathrm{v}, 3 \mathrm{~m} / \mathrm{a}\)., only \(15 /\) - each. Also RECTIFIERS, J50 (new, exGovt.), \(7 / 6\) each. Westinghouse 36 EHT 35. \(17 / 4\); 36 EHT \(100,26 / 6\) SLIDER POTS. As used in all the lacest \(T / V\) sers. A bank of four, comprising 2 or \(10 \mathrm{~K}, 100\) ohm and 500 ohm. Only \(6 /-\) the see. Easily split up.
MIDGET .0005 mfd . TWO-GANG TUNING CONDENSER, Size only \(2 \frac{z i n}{} \mathrm{i} . \times 1 \frac{1}{i n} \times 1 \frac{1}{2} \mathrm{in}\). Capacity guaranteed, standard length tin. spindle, complete with mounting bracket, less trimmers, 6/6, or complete with "built-in "trimmers, \(7 / 6\). Each plus \(6 d\), post.
RECEIVER TYPE 21. The receiver portion of the W/S 21 operating from 4.2-7.5 Mc/s. Double superhet from \(18-30 \mathrm{Mc} / \mathrm{s}\). Incorporatin B.F.O. and crash limiter. Valve line-up 7-ARP12 (VP23) and 2-AR8 (H23DD). Absolutely brand new, complete with circuit. Only 45/compleze. Vitrator Power Unit for above, brand new, 17/6 only.
I.F. TRANSFORMERS. Manufacturer's surplus. Iron-cored \(4 j 5 \mathrm{kc} / \mathrm{s}\). Size 4 in . \(\times 1 \frac{1}{8} \mathrm{in} . \times 1 \frac{1}{2} \mathrm{in}\). Per pair \(8 / 6 \mathrm{whilst}\) they last! FREQUENCY CONTROL CRYSTALS. By American G.E. Co.Oetal base fixing. Following frequencies only: \(2,500 \mathrm{kc} / \mathrm{s} .3,500 \mathrm{ke} / \mathrm{s}\), \(4,600 \mathrm{kc} / \mathrm{s}, 6,200 \mathrm{kc} / \mathrm{s} ., 8,000 \mathrm{kc} / \mathrm{s}\).-ar only each \(7 / 6\). New Condition SPECIAL COIL PACX OFFER. Limized quantity of brand new manufacturers' surplus, 3-wave-band, superhet coil packs. Iron cored. Size \(4 \mathrm{in} . \times 3 \frac{1}{4} \mathrm{in} . \times 2 \mathrm{in}\). deep. Complete with circuit, a bargain at \(25 /-\) only.
SPECIAL VALUE IN MAINS TRANSFORMERS. 250-0-250, \(90 \mathrm{~m} / \mathrm{a} 6.3\) v. 3 a., 5 v. 2 a., half shrouded, drop through type. Electrostatic screen. Price \(15 /\) - only, plus 9 d , post. Limited quantity.
RECEIVER TYPE 25. The receiver portion of the T/R II96. Covers 4.3-6.7 Mc/s., and makes an ideal basis for an all-wave receiver, as per "Praccical Wireless," August, 1949 issue. Complete with valves types EF36(2) EF39(2), EK52 and EBC33. Supplied complece with necessary conversion data for home use. Only 22/6. Chassis only, 8/6. \(6 \frac{1}{2}\) in. WAFER SPEAKERS. Special offer of the above well-known speaker, listed at 25/-. Limited quaneity avallable, brand new, at 15/. each, Post Free.
931A. PHOTO-CELL MULTIPLIER AMPLIFIER UNIT COM PLETE. Incorporating 931A photo-cell, 2 valves type 6AC7, 6 AG7, etc., etc., Can be adapted for use in window lighting, warning systems locaring foreign bodies in liquids, flaws in textiles, burglar alarms, circuit switching by relays, etc., etc. Panel size \(9 \frac{1}{6} i n . \times 3 \frac{1}{4} \mathrm{in}\). Circuit diagram not ava ilable. Price 45/- complete, post free.
R. 3515 I.F. STRIP. A complete I.F. Unit, comprising 6 SP6II.F. Stages, tuned to \(13.5 \mathrm{Mc} / \mathrm{s}\)., I EA50 diode detector, and I EF36 or EF39 output or video stage. A few modifications only are required to adapt this unit, which will give pictures of extremely good quality. Price, complete with valves, and foolproof modification instructions, is \(45 /\)., plus 5/-carriage and packing. Limited quantity only.

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R.C.A. ET-4336B COMMUNICATIONS TRANS.

MITTER, 2-20 mcs. Height 57in., width 17in., depth \(24 i n ., 9\) valves: \(4 / 866,2 / 805,2 / 813,1 / 807\). New. Comprehensive booklet and photographs available on loan against deposit of El returnable. Crated and carriage paid, 560 each.
a MAX B4/40 TRANSMITTER. Listed 475 . Shopsoiled oniy \(\mathbf{6 5 0}\). Full details on request.
AR88 LF. Equal to brand-new, £45. Crated and carr. paid.
1155 RECEIVERS. Brand-new, in transit cases, complete with valves. Without a blemish. Aerial tested, 69/10/ -, plus \(7 / 6\) carriage.
R107. The weil-known Army Communications receiver. Few only, \(112 / 10 /\)-, carr. paid.
CANADIAN NO. 58 TRANS/RECEIVERS, 33-50 metres, complete with mikes, phones and 2 batteries (1 spare). Each complete unit, 10 gns. Guaranteed brand-new.
1155 POWER PACK AND OUTPUT STAGE ( U 50 and KT61) to "Wireless Worid" specification, in neat black crackle case size \(12 \mathrm{in} . \times 8 \mathrm{in}\). \(\times 5 \mathrm{in}\)., with 5 in . L.S. bullt in. Just plug into 1155 and set is instantly all A.C. operated without any modifications. Price \(\mathbf{5 5 / 1 9 / 6}\), plus 3/6 carriage.
TANNOY AMPLIFIERS. Rack type audio amplifiers fitted with 10 valves: \(4 / 163,4 /\) KT \(66,2 / \mathrm{U} 18\). Used ex-Govt., but fine condition, \(£ 6 / 10 /\). Without valves, \(£ 3\). 6 ft . rack free of charge to purchasers of 4 amplifiers with valves. This is a unique offer.
TELEVISION PRE-AMPLIFIERS. Makes a blur into a clear picture. Really sensitive. Fitted with EF50. Very compact. Power requirements 6-3 v. L.T. and 200/250 H.T. Ready for instant use. No modifications. State for London or Birmingham. Price \(15 /-\), plus \(1 /=\) post and packing. A snip.
TU6B Tuning Units, 3000, \(4500 \mathrm{kc} / \mathrm{s}\), less outer case. New condition, \(10 /-\), carriage \(1 / 6\).
BC306 Aerial Tuning Units, ceramic switch, slowmotion drive, etc. Size \(16 \mathrm{in} . \times 8 \mathrm{in} . \times 8 \mathrm{in}\). Brand-new in cartons. The last few dozen. 10/- each, carriage \(1 / 6\). 12-volt Car Radio. Comprises an ex-Govt. 6 -valve Command receiver expertly modified to tune medium wave band. Complete with separate Sin. loudspeaker. Excellent performance (definitely not available for 6 volts), price \(£ 5 / 10 /-\), carr. paid.
P.M. Loudspeakers. Best makes : 8 in ., \(13 / 6\), post \(1 /=\); 12 in., \(39 / 6\), postage \(1 / 6\).
PAMPHONIC PA LOUDSPEAKERS. 10in. high flux unit. Handsome, Maroon Cellulose, Metal Cabinet, \(20 \times 9 \times 13 \mathrm{in}\). Impedance 3 ohms. Brand-new. In cartons. Not surplus. Less than half price, 55/-, carr. paid. Command Receiver Triple Controllers. Brand-new, in cartons, \(8 / 6\), post paid.
Medium Wave Coils for converting BC453/4/5 to MW., 10/- each. State which required. Tuning spindle and knob for same, 2/6. EF50 ceramic valve hoiders, 5!- doz. (minimum). Morse keys, new, ex-Govt., brass, fine job, 3/6, post 6d.
AMPLION CRYSTAL PICK-UPS, brand-new, in cartons, 26/9, post paid. Type 25 receiver, with 6-6 v. valves, 22/6, carr. paid. NEW.
CONDENSERS, waxed tubular. 50 for \(10 /-, 25\) of .11000 v . and 25 of .25 mfd .500 v . Lexington Junior Moving Coil Pick-up. One only 70\%, post paid. All offers subject to goods being unsold. Above are precise and honest descriptions. Enquiries must be accompanied by \(2 \frac{1}{2} \mathrm{~d}\). stamp.

\section*{H.P. RADIO SERVICES LTD. \\ Britain's Leading Radio Mall Order House Estd. 1935}

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\section*{COMMUNICATIONS RECEIVER TYPE} RII55.
This famous receiver needs no description, but for those not familiar with the set we can supply full circuit details, etc., for 9d, post free. All sets are air-tested before despatch and are supplied complete with 10 valves. We will gladly demonstrate any ses to callers. Brand new. t \(9 / 10 / 0\), or desparched in maker's transit case for \(£ 9 / 19 / 6\), carriage paid. A.C. mains power pack for \(£ 9 / 19 / 6\), carriage paid. A.C. mains power pack
and output stage for RIIS5 receiver enables and output stage for RIIS5 receiver enables
this receiver to be operated direct from A.C. this recelver to be operated direct from A.C.
mains. Just plug in! Price \(£ 3 / 10 / 0\), carriage paid.

\section*{MEDIUM WAVE COMMAND RECEIVER.} Frequency range 200-600 metres. Valve line up: \(125 K 7\) R.F., \(12 K 8\) F.C., \(12 S K 7\) I.F., \(125 K 7\) I.F., I2SR7 Det. B.F.O. and 12A6 output (all metal valves). Needs only:50 k. pot. for volume control purposes to be fitted and ten minutes work and receiver is all ready for operation as either a car radio or domestic radio. Note: this receiver is the original medium-wave set with the correct I.F. for she job and not a bodgedup conversion. Unused and in excellent condition, only \(£ 3 / 19 / 6\). Circuit details supplied free with set.

\section*{RECEIVER TYPE 18.}

A four valve battery superhet receiver. Uses standard iron-cored \(465 \mathrm{k} / \mathrm{cs}\). I.F. transformers. Complete with all valves. Frequency range \(6-9 \mathrm{mc} / \mathrm{s}\). Circuit diagram provided. Brand new. Only \(17 / 6\).

\section*{RECEIVER TYPE II47B.}

A 7 -valve UHF receiver, range approx. 200 megs. Particularly suitable for conversion to \(144 \mathrm{mc} / \mathrm{s}\). Valve line-up : two EF50, two EF36 and one each of EBC33, RL7, RLI6. This receiver is beautifully constructed and fitted with micro-condenser drive. Contained in black metal case size \(8 \times 7 \times\) Gin. BRAND NEW IN USEFUL TRANSIT CASE which might have been designed as a tool box! ONLY 30/-, carriage paid.

\section*{RECEIVER UNIT 25/73 (TRII96).}

Valve line-up : two EF36, two EF39, one EK32 and an EBC33. Easily and rapidly converted to a very fine superhet receiver. (See "Practical Wireless" August 1949.) Supplied complete with circuit and conversion data. In very good condition indeed and offered at 22/6, post paid.

\section*{RECEIVER UNIT TYPE 6.}

Housed in case size \(10 \times 8 \times 7 \mathrm{in}\). Frequency range \(3-7 \mathrm{mc} / \mathrm{s}\). Valve line-up:-five of EF50, and one each of EBC33, and EF36. The dial is exceptionally well geared and calibrated. Slightly soiled condition externally but perfect inside 2916. Callers only.

RI355 RECEIVER.
New, but slightly store soiled. 49/6, plus 5/carriage.

\section*{45 MC/S PYE STRIP.}

A ready made vision receiver! (London frequency). Uses six EF50 and one EA50. Complete circuit data provided. All brand new. Less valves, \(39 / 6\), with valves \(62 / 6\).

MODULATOR TYPE 67.
This unit is essentially a heavy duty mains power pack for 230 volts A.C., 50 cps. operation. It contains a heavy duty mains eransformer with two 6.3 volt outputs, 5 voles and the H.T. is 345-0-345 volts. Heavy duty choke appropriate smoothing condensers, rectifier, etc. In addition there are sundry SP6|'s, ete., which we haven't listed. Don't miss this bargain at only \(52 / 6\), plus \(5 /\)-carriage.

WIRELESS SET NO. 48. Tranismitter/ Receiver.
Brand new ex-U.S.A. equipment. Frequency coverage \(6-9 \mathrm{mc} / \mathrm{s}\). A complete station. All brand new in six cartons. Comprises 1 transmitter/receiver type 48 , complete in case, with ten valves and one \(1-\mathrm{mc} / \mathrm{s}\). crystal. Two sacchels, one ground aerial, spare valve case, complete with ren spare valves, one key assembly, one hand microphone, two headsets, one hand generator together with mounting tripod, ewo battery boxes and all necessary interconnecting cables and comprehensive instruction manual. Special offer for one month only at less than serap price. Only \(£ 9 / 19 / 6\), plus \(10 /\) carriage. Inland only. Extra for Eire and Overseas.

BRAND NEW AMERICAN MADE 19 SETS.
This is a transmitter/receiver covering 2-8 \(\mathrm{mc} / \mathrm{s}\). on phone, C/W. and MC/W. It also comprises UHF transmitter/receiver and an intercomm. amplifier. Supplied complete with 15 values. Only Ei5, plus 10/- carriage.

\section*{INDICATOR UNIT II6H.}

Brand new and contained in manufacturers original wooden crates. A more up-to-date version of the famous 6A. The contents include VCR97 tube, four EF50, three EB34 and innumerable other useful components. Price only 7916, plus \(7 / 6\) carriage and packing.

\section*{RECEIVER TYPE 21.}

A battery operated superhet receiver covering 4.2-7.5 Mes. and 19-31 Mcs. Operates as a double superhet on the 10 metre band. Complete with hine 2 v . valves. Cireuit diagram supplied. In hew condition. 45/- carriage paid.

\section*{6 V. VIBRATOR PACK.}

Suitable for use with the above set. With circuit diagram. Output 150 v . at 40 mA fully smoothed and rectified. Only \(17 / 6\) post free.

\section*{ADMIRALTY RECEIVER B36 OR B2I.}

A superb communications receiver made by Marconi's Wireless Telegraph Company. Valve line-up: two R.F. stages, freq. changer, three I.F. stages, double diode triode, BFO and output stages, doube diode triode, i-20 Mc/s. ( \(15-300\) metres) in 4 unbroken bands via turret coil change. metres) in 4 unbroken bands via turret coirchange. Sensitivity for 10 dbs signal to noise ratio-better Sensitivity for 10 dbs signal to noise ratio-better
than I mierovolt! Filament cransformar for 230 than IC mierovit operation is included but the H.T. supply is v. AC operation is included but the H.T. supply is
required. Necessary valves are: International required. Necessary valves are : International
Octal types two of KTW61, two of X65, three of Octal types two of KTW61, two of X65, three of
KTW63, one of DH63 and a KT63. In new and KTW63, one of DH63 and a unused condition. Complete with circuir diagram. Supplied less valves. \(\mathrm{f} 12 / 10 /\). carriage paid. Only 20 of these high grade receivers in stock. First come first served!

INDICATOR UNIT 62A.
Contains twelve EF50, four of SP61 (CV||8), two of EB34, three of EA50, 13 pors and a VCR97 sube. The finest Indicator Unit of them all. Unused condition but chassis slighty soiled. Offered to callers only. Price \(84 / 19 / 6\).

\section*{LOUDSPEAKERS,}

Plessey 10 in . p.m. type. ALNI magnet. Suitable for T.V. Brand new and boxed. Only 17/6 post paid.

\section*{MICROPHONES.}
M.C. microphone by Vitavox. With on/off switch. 30 ohm impedance. Brand new and boxed. 30/- post paid.

LOOK ! Marconi/Ekco Signal Generators cype TFII4G. Marconi/Ekco valve voltmeters. Call for details.

HEAVY DUTY 6 V . ACCUMULATORS. Size \(12 \times 9 \times 7 \mathrm{in}\). Brand new. Caparity 85 ampere hours. In teak case. Only \(59 / 6\). Exceprional value.

\section*{METAL RECTIFIERS.}

Selenium full wave bridge 12 v. 5 amp., 15/- each.

\section*{MAINS TRANSFORMERS.}
\(350-0-350\) volts \(80 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v}\). tapped 4 v . at 3 amps., 5 v . tapped 4 v . at 2 amps. Top chassis mounting. Fully guaranteed. 18/6, plus 9d. postage.
\(250-0-250\) voles, \(130 \mathrm{~m} / \mathrm{a} ., 6.3 \mathrm{v}\). at 6 amps. . 5 v. at 3 amps. Fully shrouded top chassis mounting. A quality job. Guaranteed. 27/6.
E.H.T. Transiormer for VCR97 or VCR5I7 tubes. 2,000 volts, 4 v . for cube heater and 4 v . capped at 2 v. for E.H.T. rectifler. \(30 /\)-.
All the above transformers have standard primaries, \(200 / 250\) voles 50 cycles.

\section*{SPECIAL OFFER}

Mansbridge condensers. 4 mid. 750 v. wkg. All brand new. Six for 10/6. post pald.

SPECIAL BARGAIN FOR CALLERS. UHF RECEIVER TYPE I4BI.
Frequency range \(65-86 \mathrm{mc} / \mathrm{s}, 6\) in. slow motion drive, II of 6.3 v . valves as follows: four of EF39, one of EK32, four of SP61, one of P61, one of EB34. I.F. is \(12 \mathrm{mc} / \mathrm{s}\), B.F.O. etc. Brand new with circuit. \(£ 3 / 10 / 0\).

Full range of valves, \(C\) R. tubes, components, etc., at very attractive prices. Many, many bargains for the caller. Large quantities of resistors, condensers, valves, etc.s for EXPORT AND TRADE DISPOSAL.

THE " EASYBUILT" TELEVISOR
COMPLETE DETAILS OF A FULL SIZE MAGNETIC TELEVISOR IN BOOK FORM. DETAILED INSTRUCTIONS, WIRING DIAGRAMS, ETC., A MINE OF INFORMATION FOR ONLY 2/6, POST FREE.
Inexpensive Televiston data booklet No. 4. Fully revised and brought up to date. Deseribes the \(45 \mathrm{~m} / \mathrm{cs}\). "Pye "strip as well as the R1355 and other well-known government surplus units. Only \(2 / 6\), post free.


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II, UPPER SAINT MARTIN'S LANE, LO NDON, W.C. 2

TEM O545
3 minutes from Leicester Square Station (up Cranbourne Street) Shop Hours : 9-6 p.m. (9-1 p.m. Thursday). Open all day Saturday


\section*{FROM EVERY POINT OF VIEW -A VAST IMPROVEMENT}

The Magnavista Lens is not just a " magnifier "; it is an optical instrument, designed by optical scientists. For perfect clarity, and to avoid distortion, different receivers need different lenses. There is a Magnavista Model specially designed for your set. Here is real television progress which you can see for yourself.

\section*{PRICES:}

There are over 18 different Magnayista models at prices ranging from c4-14-6 to 47 -17-6.


\section*{NEW}

\section*{R.C.A. TELEGRAPH AND TELEPHONE COMMUNICATIONS TRANSNITTERS TYPE ET. 4336.}

Designed and produced by the Radio Corporation of America, the Type ET.4336 is outstanding in the exceptional quallty of construction, versatile operation with rapid band selection and adjustment, and the embodiment of modern advartageous features not generally found together in one equlpment.
The transm tter is designed for communications service in The transm ter ig designed for communications service in the high frequency range of 2,200 to 20,000 kilocycles, and any desired operating frequency whin selected, using ordinary crystals. All controls quickly selected, using ordinary crystals. All controls employed in tuning the transmitter to any frequency within
the overall range are mounted upon the frant panel. Shift the overall range are mounted upon the front panel. Shift
from telegraph to telephone operation is switch controlled, and high speed keying is permissible. Two transmission power levels of 250 or 350 watts are available, however the actual outputs obtained are somewhat in excess of these values at \(20 \mathrm{mc} / \mathrm{s}\), and increase with decreas ng frequency to values in excess of 300 and 450 watts respectively at \(3 \mathrm{mc} / \mathrm{s}\).

\section*{GENERAL FEATURES.}

The complete transmitter is housed in a tall console cabinet, superbly finished, and has a very attractive appearance. Side and rear panels are removeable, and electrical interlocks are fitted as a safety feature. Stylish panel controls are conveniently grouped, and clearly identified. Five Meters are employed to read Aerial, Plate and Grid currents, and Filament voltages. A modulation and keying indicator of the vapour column type is mounted on the front panel. The Type M1-19468 Crystal Multiplier, which we supply with the transmitter, slides into an aperture which is normally concealed by a removeable panel. When employing this unit, the oscillator stage in the transmitter functions as an Intermediate P.A. stage, or frequency multiplier, and whilst employment is optional, transmission over the entire transmitter frequency range, using low frequency crystals, is a distinct operational advantage.

\section*{CIRCUITRY.}

Valves Type 807 are employed in the Crystal Multiplier Unit and the Transmitter Crystal Controlled Oscillator. The Power Amplifier stage utilizes two Valves Type 813 operating in parallel, and the Modulator stage uses two Valves Type 805 operated in a Class B arrangement. Plate and screen voltages for all stages are obtained from a mercury-vapour rectifier comprising four Valves Type 866A, connected in a full-wave parallel circuit. An Antenna Coupling and Matching system is provided, and is so designed to allow the use of various feeder arrangements.

\section*{ELECTRICAL CHARACTERISTICS.}

Type of Modulation
Audio-Input Impedance
Audio-Input Level for 100
per cent. Modulation
A-F Resoonse
A-F Harmonic Distortion
Power Input
Telegraph, Low Power
Telegraph, High Power
Telephone, 100 per cent Modulation
Power Supply Require ments

Regulation (Maximum)

Class B, high level.
500 ohms.
20 vu.
\(\pm 5\) db from 400 to 7,500 cycles 5 per cent r.m.s.
1.25 kW .
1.46 kW .
1.82 kW .

115-230 v., 1 phase, 50-60 cycles. 5 per cent.

DIMENSIONS. Height, 58in. Width, 17 in . Depth, 24 in . WEIGHT. 500 lb

CONDITION. New and perfect, as ex-factory. Our Guarantee and Individual Test Certificate is supplied with each transmitter.

We offer immediate dellvery of the above equipment to home and overseas buyers; complete and with all valves, at an exceptionally low price.
Full specification and further particulars are promptly avallable on request.

\section*{LAWRENCES}

6I, BYROM STREET, LIVERPOOL, 3.
Telephone: CENtral 4430.
Experienced Export Shippers.
A GUARANTEE OF SATISFACTION WITH EVERYTHING WE SELL

Moving Collspeakern. All wew \(2 / 3\) ohms P. M.: \(3 \mathrm{ib} \cdot, 11 / 6\) Rola 166 ; 4 in . 18/-; 5in. \(10 /=\); (with Trapa, 14/6) : biln. \(11 / 6\); 8in. 12,6 and \(17 /-\) ( with Trans. 16 -)

Potentiometers. New Centralab: \(2 \mathrm{~K}, 5 \mathrm{~F}, 10 \mathrm{~K}, 23 \mathrm{~K}\), \(50 \mathrm{~K}, 100 \mathrm{~K}, \frac{1}{}, 1\) and 2 meg., less switch, \(3 / 9\), with t. 1 and 1 meg., \(6 /\)-.

Surplus Potentiometers. All standard aize with extended spindles: \(5 \mathrm{~K}, 10 \mathrm{~K}, 15 \mathrm{~K}, 20 \mathrm{~K}, ~ 35 \mathrm{~K}, ~ 50 \mathrm{~K}, 100 \mathrm{~K}\),
250 K j00K, \(750 \mathrm{~K}, 1\) meg. \(2 / 6 \mathrm{ak} \mathrm{h}\).
Coils. Denco Maxi" \(Q^{\text {" }}\) Bigh " \(Q\) " with miniature size Iron Duat Cores. Aeriai, II.F., or Oscillator for \(46 \mathrm{sk} \mathrm{k} / \mathrm{c}\) or \(1.6 \mathrm{~m} / \mathrm{c}\). Bange covers all uavebauds from 3.6 to 2.000 metres. Wiring diagram. Pri ces : Chasale Mount\(\mathrm{lng}, 3 / 9\) (Nith React. 4/9) Octal Atting pln base,
\(4 /-\) (uith Renct., \(5 /-\) ). Denco T. R. F. Matehed, pair Medimand Long Wavea, B/6 palr. Weymouth T. R.F. Matched pair M. and L. Waven, \(9 / 6\) pair. Superhet
Mitched pair 8 . M. and L. Waves, \(8 / 9\) or \(11 / \AA\) pair. Mitched pair 8. M, and L. Waves, \(8 / 9\) or 11/A pair. Allyper Woanite
mouth Midget 1 in . \(\times\) dian., Iron Oore. Aeribi. K.F.
or Osc., \(3 / 6\) each.

Electrolytic Condensers. B.E.C. Midget Can Tubular, 8 mfd .450 v . (14in. \(\times \mathrm{fin}\). dia.), \(2 / 6 ; 8-8 \mathrm{mfd} .450 \mathrm{v}\).

 (1/in. \(\times 1 / \mathrm{in}\) dia.), \(4 / 6 ;\) Dubilier "Drlitic" Card \(16 \mathrm{mld} .500 \mathrm{v} .5 / 6\); B.I. Can, standard slze, 8 mfd . 500 ₹., \(3 / 9 ; 8-8 \mathrm{midd} .500\) ₹., \(5 /-\); 16 mfld .500 v., All New Stock. Dabilier mich. \(001 \mathrm{mfd} .5,000\) v.,2, 6 .
Denco IF. Liner for accurately lining-up \(465 \mathrm{k} / \mathrm{c}\), or \(1.6 \mathrm{~m} / \mathrm{ci} . \mathrm{F}\). channels. Pre-tupe circuita, battery opera-
ted and completely self-contalued. Price \(42 /\) - (incl. P.T.).

Coil Packs. Osmor Midget Coil Pack. Size \(31 \mathrm{fu} . \times 2 \mathrm{fa} . \times\) lin. covering B, M, and Lh. Waves. Colls wound on Polyatyrene Formers with adjuatable Iron Cores
ensures efficient performance. Factory wined and aligned. Price, Including full clrcuits for saperhet 4Kik/c. Unit, \(33 /\)-. Plus \(7 / 4 \mathrm{P}\). T. Weym outh Midzet 3 inn. \(\times 2\) in. \(\times 1 /\) in. coverngg 8-M-L W/Bands, ior W/Band, \(42 / 10\). Wesrite Iuperhet Type 705 and 706 , Wize 4 ing \(x\) sgin. \(x\) ilin. covering two short Wave and one Medium W/Bavd or A.is-L Wavebands, \(465 \mathrm{k} / \mathrm{c}\), employs Iron Cored Coils 37110 . All of the above Coil Packs include 8witchlug, Padding and
Trimuez Condensers.


Outpak Transformer-8tern's, Midget11in. \(\times\) 3in. \(\times 1 \mathrm{in}\) ratio 60-1, \(3 / 9\) (or ratio 90-1, 3/9). Stern's Mult Stern's Heary Daty Multi-ratio. aflot 12 Tapped handle 13 wattu and suits P.X.4s, 6L6s, etc., 28/6. Rola Multi-ratlo. \(5 / 6\) watts, \(10 / 6\).
L.F. Chokes. Midget 10 henry 250 obm \(40 \mathrm{~mA} .4 / 6\)
 20 hay, 250 olim \(120 \mathrm{~mA} ., 18 / 6 ; 9\) hay. 250 ohm
\(150 \mathrm{~mA} ., 8 / 6\). Aluminiam Chassls. Subatantially made of gauge 16
S.W.G. with four sides, \(7 \mathrm{~km}, \times 4 \mathrm{in}, \times 2 \mathrm{~m} .3 / 3 ; 3 \mathrm{Ha} . \times\) \(\operatorname{in} . x 2 / \ln .41-; 101 \mathrm{n} . \times 6 \mathrm{in} . \times 2 \ln .4 / 11 ; 10 \mathrm{in} . x\)
 \(16 \operatorname{in} \times 8\) in. \(x 310 ., 8 / 6\).
1.F. Transtormer, 485 l/a- New well-known manufac-
 \(3 \ell \mathrm{in} ., 16 / 6\) pair. Weurite Stand Cap, Iuned, \(465 \mathrm{k} / \mathrm{c}\)., \(20 /-\mathrm{pr}\). New Surplus \(465 \mathrm{k} / \mathrm{c}\), Iron Core, \(4 \mathrm{in} \times 1 \frac{1}{2} \mathrm{in}\). sq-, 10/= pair.

Meter Re tiffers. Westinghouse 250 micro,amp., 11/6; \(1 \mathrm{~mA}, 10 / 6 ; 5 \mathrm{~mA}, 4 / 9\).
 \(5 / 6 ; 250\) ₹. \(100 \mathrm{~mA}, 7 / 6 ; 250 \mathrm{~F} .170 \mathrm{~mA}, 13 / 8\).
F. Wane Bridge Eectifiers for Battery Charging or Models 2, \(40: 6 \mathrm{v}, 11 \mathrm{amF}, 7 / 6 ; 6\) or 12 v . \(1 \frac{1}{\mathrm{smmp}} \mathrm{mm}, 11 / 6\); 12 or 24 ч. 3 amps, \(23 /-\) Tatable Resistor to control charging (or model speed
up to 3 amparatirg, 13/6. Suitatle METER, \(5 / 9\).

Charger Transformers. Sultable for use with above Rectifiers. Each has Lapit of 830 volta. Outputs (a) 24 roits tapped 16 v., 9 . gad 4 \%. at 3 amps.,
\(22 / 9\); (b) 30 volta tapoed 25 v. and 9 . At 9 ampa 2319 ; (c) 15 volta tapped 9 v. and ampa, 16,6 ; (d) 6 yolts. 11 a mps., \(11 / 3\) : (e) 16 voits tapped 18 st A Battery Charger Wiriag Diagram la included fith purcbase of Charger Transformerand flectifier.
NEW SURPLUS ! ! FOSTER TRANSFORMERS. PRIMARY 230 VOLTS, SECONDARE 12 VOLTS, AMPS. 21
Filament Transtormer. Inputs 230 volts, outputs \(6.3 \mathrm{\nabla}\). 11 amp., \(8 / 3: 4\) V \(^{2}\) 11 amp.. \(7 / 6\); 1nput \(200 / 250\) v.
 Westinghouse Reotifers. (a) F.T.s1, Lated 350-0-300 rolts \(100 \mathrm{~m} / \mathrm{a} .35 /-\); (b) H.T.52. Rated \(350 \cdot 0-350 \mathrm{v}\)., \(200 \mathrm{ma} / \mathrm{n}, 37 / 6\) : (c) H.T.63. Rated \(500-0-500\) จ., \(200 \mathrm{~m} / \mathrm{f} ., 50\) - ; (d) Type \(16 \mathrm{H} . \mathrm{T} .56,17 / 9\); (e) Tyne \(1 \mathrm{~m} / \mathrm{a} ., 17 / \mathrm{B}^{\prime}\); (g) Type 36E.K.T.100, Pulse rating \(1 \mathrm{ma} / \mathrm{B} ., 17 / 6\)
\(5 / 6 \mathrm{~V} . .26 / 8\).
Mains Transformers, All New Etock with Primarles tapped for 200-250 volts. Secondaries: (a) 250-0-250 (Tapped 4 v.) 2 smp., 18 ' 6 (also avallable with 350 -0. 350 volt at \(18 / 6\) ). (b) Stern's \(350-0-350\) volt, \(150 \mathrm{~m} / \mathrm{a}\). 6.3 v. (Tapped 4 v.) 4 amp. and \(8 \mathrm{v}. \mathrm{(Tapped} 4 \mathrm{v}\).) \(2 \mathrm{amp} ., 3816 ;\) (b) \(850-0-350 \mathrm{~V} .250 \mathrm{~m} / \mathrm{a}\)., 4 voit 8 amp., 45.3 nnup. 6.3 v . (Tapped 2 v .), 2 amp, and
6.3 v .6 amp. 72/6, and many to other ratings.
Heater Auto Transformers : (a) Tapped \(2 \nabla\)., 4 v., 5 \(\nabla\)., and 6.3 volts 3 amp., \(9 / 6\); (b) 4 v. 3 amp. to 8 V .
2 amp. Reversthle \(6 /=\); (c) 4 v. 3 amp. to \(6.3 \nabla .2 \mathrm{amp}\). Reverslble, \(6 j-\).

Power Potentiometers - Bugin aijuatable slder type max. 6 ohma 60 watt, or 14 ohms 60 watt, \(13 / 6\) each
Ex-Govt. Rotary Adjustable 50 ohms 60 Fatt \(5 / 9\).

A Station Preset Tuner. A compiete self-contaimed unit from which any 3 Medium W/Band und 1 Long W/Band stations may be pre-selected, and then Tuning Condenser required only 4 connectlona are tecessary, price 40/\%.
 \(15 \mathrm{mmp}, 1500\) ohms, \(5 /-;(\mathrm{B}) .2 \mathrm{amp}-, 1.000\) ohms,
\(4 / 3 ;(0) .3\) amp. 600 olums. \(5 /=\).

6 Volt Vibrator Transformers, secondary \(350-0-350\) volts, \(85 \mathrm{~m} / \mathrm{h} ., 8 / 6\).

\section*{KITS OF PARTS AND CONSTRUCTORS ASSEMBLY OUTFITS}

\section*{"6 MAINS OF BATTERY PERSONAX EIT '}

A complete KIT OF PARTS to buidd our new MidaET 4-VALVE SUPERFLET or BAi TER Y operationis now avallable.
Thls 2 Waveband Buperhet Receiver is deelgned to operate on A.C. mains 200/240 volts, or by an "All-Dry" Battery, elther method being selected by meana of a Rotary suitch. It is so dusigned that the Maibs Section (size 4 in. \(\times 1\) in. \(\times 3\) in.) ly surplied as a saparate kit, which may be added at ary time. The kit therefore can be
supplied elther as an "All-Dry" Battery fersonal set, of as a Midget Recelver for combined Battery/Mains operatiou

The citcuit incorperateq Delayed A.V.C. and Preaelective Audto Feedback. A Rola 4in. P.M. speaker with generous she Output Transformer ensures excellent quality reprod.ction. Ready Wound Frame Aerlal, Fully aligned I.F. Tranaformerb, and a
Drilled Midget Chasmis are lncluded. Valve line up :-T.R.5 (F.C.), 1.T.4 (I.F. Amp.) Drilled Midget Chaspis are lncluded. Valve line up :- - .R.
The fet is quite easily bullt fiom the sery detafled assembly instructions supplied, hach include a p. actical component layout, uith potnt to point wiriug, and a circuit diagram for both the set and the Mains Unit.
Price of COMPLFTE KIT (less M 3 ins Unit), \(26 / 13 / 9\). Prlce of COMPI.ETE MAINE UN1T KIT, £1/17/6. EVER READY Type B114 BATTERY, \(9 / 7\).
An attractive Walnut Finlaned Cabinet, size 9in. \(\times 5+\mathrm{im} . \times 5 / \mathrm{ln}\)., of the hinged lid type ani sultable to house the combined set is available for \(19 / 9\).

THE COMPLFTE AGEFMBLT JNSTRUCTIONS mentioned above can be suppplied se arately for \(1 / 9\).
 Midget 4 -talye All-iry Battery Personal Set. Conaiata of Regenerative T.R.P. Circuitempl Jying in st Cuned Frame Aerial, with Denco iron Dust Cored Conl, thereby

Valve Iine-up: IT4 (R.F. Ampl.), IT4 (Detector), \(1 S 5\) (1st A.F.), and 3S4 (output) Includes iatent Rola 3in. Moving Coil Speaker, and a Charsis already drilled and for a cablmet, minimum size \(6+\ln\). \(x\) tifn. \(x\) 3in. Detaijed Building Instructions, with Practical Layout and Clrcutt included with Ktt make assembly easy. Prict for C smplote Kit, \(£ 318 / 9\) (pius \(16 / 7\) P.T.). Sultable unpolished Cabinet \(61 \mathrm{ln} . \times 4 \frac{\mathrm{in}}{} \mathrm{x}\) Sin. 12/9. Ever Ready B114 Battery, \(9 / 7\). Building Instructions, Circuit, ete., supplied separately, 1/-
A Complete Kit of Parts to bulld a Midset "All-Dry "Battery Ellminstor, giving sppocox. 69 vilts H.T. and L. 4 volts L.T. This Eliminator ts suitable for any Personial set requembled and is boused in a Light Aluminiun Cabe, size 4 fln . \(x\) igin. \(x\) 3in. It
* Send 3d. stamp for our Comprehensive Stock List. When ordering please cover packing and postage.
cat wherefole be sac
plete Kit, £11/17/6.
Wireless World" Midcet A.C. Mains 2-Valve Beceiver. We can supply all the comissue at a total cost of \(83 / 5 / 0\). Reprint of detailed assembly lnstructions and circult supplied separately for 9 d .

WIreless World " Midget A.C. Msins 3-Vave Receiver, Coverirg Long and Medium Wavebands. We can supply all the Components, facluding Irliled Chassis, Valvea, Moving Coil Speaker, ete., to build tnis Set. as specifled in the Fob. isave, at a total cost of \(£ 4 / 10 / 0\) including a reprint of the complete Arsembly Inetructions and Ctrcult (this is available reparately for 9 d .) and Practical Componenta Layout with pointto point" connections, An attractive Walnet Finished Cabinet ts bow available for Assembly which includes the latest Station Name Dial and Dial Encutaheon, and a Combined Switch/Volume Control to effect very slight modification. Inciusive price, Combined Switch/Volume Controith effect wery aight modification. Inciusive price,
We oan supply all the Components, including Vaives, M/Coil Speaker, etc., to build a Midget A.c.id.c. Mains T.R.F. 3-Valre (Plus Metal Rectifler) Recelver as designed and specited by a popular Tecinical hagazine, at a
the assembly tartructions, and layout avalabie for 9 d .

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[^4]:    American Journal of Mathematics, 1879, and Phil. Mag., R80, Series 5 , Vol. 9, p. 225.

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