

15-240 Watts!

HY5 **Preamplifier**

The HY5 is a mono hybrid amplifier ideally suited for all applications. All common input functions (mag Cartridge, tuner, etc) are catered for internally. The desired function is achieved either by a multi-way switch or direct connection to the appropriate pins. The internal volume and tone circuits merely require connecting to external potentiometers (not included). The HY5 is compatible with all LLP, power amplifiers and power supplies. To ease construction and mounting a P.C. connector is supplied with each pre-amplifier. FEATURES: Complete pre-amplifier in single pack—Multi-function equalization—Low noise—Low distortion—High overload—Two simply combined for stereo.

APPLICATIONS: Hi-Fi-Mixers-Disco-Guitar and Organ-Public address

APPLICATIONS: Hi-H—MIXERS—DISCO—Guntar and Organ—rabite address SPECIFICATIONS:
INPUTS. Magnetic Pick-up 3mV; Ceramic Pick-up 30mV; Tuner 100mV; Microphone 10mV; Auxiliary 3-100mV; input impedance 4.7kΩ at 1kHz.
OUTPUTS. Tape 100mV; Main output 500mV R.M.S.
ACTIVE TONE CONTROLS. Treble ± 12dB at 10kHz; Bass ± at 100Hz.
DISTORTION. 0-1% at 1kHz. Signal/Noise Ratio 68dB.
OVERLOAD. 38dB on Magnetic Pick-up. SUPPLY VOLTAGE ± 16-50V.

Price £6 27 + 78p VAT P&P free.

HY30

15 Watts into 8Ω

The HY30 is an exciting New kit from I.L.P. It features a virtually indestructible I.C. with short circuit and thermal protection. The kit consists of I.C., heatsink, P.C. board. 4 resistors, 6 capacitors, mounting kit, together with easy to follow construction and operating instructions. This amplifier is ideally suited to the beginner in audio who wishes to use the most up-to-date technology available

FEATURES: Complete Kit-Low Distortion-Short, Open and Thermal Protection-Easy to

APPLICATIONS: Updating audio equipment—Guitar practice amplifier—Test amplifier—

SPECIFICATIONS: OUTPUT POWER 15W R.M.S. into 8Ω : DISTORTION 0·1% at 1.5W. INPUT SENSITIVITY 500mV. FREQUENCY RESPONSE 10Hz-16kHz-3dB. SUPPLY VOLTAGE \pm 18V.

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HY50

25 Watts into 8Ω

The HYSO leads I.L.P.'s total integration approach to power amplifier design. The amplifier features an integral heatsink together with the simplicity of no external components. During the past three years the amplifier has been refined to the extent that it must be one of the most reliable and robust High Fidelity modules in the World.

FEATURES: Low Distortion—Integral Heatsink—Only five connections—7 amp output transistors—No external components

APPLICATIONS: Medium Power Hi-Fi systems—Low power disco—Guitar amplifier

SPECIFICATIONS: INPUT SENSITIVITY 500mV OUTPUT POWER 25W RMS into 8 Ω LOAD IMPEDANCE 4-16 Ω DISTORTION 0·04% at 25W

af 1kHz SIGNAL/NOISE RATIO 75dB FREQUENCY RESPONSE 10Hz-45kHz—3dB. SUPPLY VOLTAGE ± 25V SIZE 105 50 25mm

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HY120

60 Watts into 8Ω

The HY120 is the baby of I.L.P.'s new high power range. Designed to meet the most exacting requirements including load line and thermal protection this amplifier sets a new standard in

FEATURES: Very low distortion—Integral heatsink—Load line protection—Thermal protection—Five connections—No external components

APPLICATIONS: Hi-Fi—High quality disco—Public address—Monitor amplifier—Guitar and organ

organ SPECIFICATIONS INPUT SENSITIVITY 500mV. OUTPUT POWER 60W RMS into 8Ω LOAD IMPEDANCE 4-16Ω DISTORTION 0·04% at 60W OUTPUT POWER 004 AND THIS OF EACH STATE OF THE STATE OF T

± 35 v SIZE 114 50 85mm

Price £19 01 + £1 52 VAT P&P free.

HY200

120 Watts into 8Ω

The HY200 now improved to give an output of 120 Watts has been designed to stand the most rugged conditions such as disco or group while still retaining true Hi-Fi performance. FEATURES: Thermal shutdown—Very low distortion—Load line protection—Integral heatsink external components

APPLICATIONS: Hi-Fi-Disco-Monitor-Power slave-Industrial-Public Address

SPECIFICATIONS
SPECIFICATIONS
INPUT SENSITIVITY 500mV
OUTPUT POWER 120W RMS into 8 \(\Omega\$ LOAD IMPEDANCE 4-16 \(\Omega\$ DISTORTION 0.05 \)% at 100W at 14Hz
SIGNAL/NOISE RATIO 96dB FREQUENCY RESPONSE 10Hz-45kHz-3dB SUPPLY VOLTAGE

±45V SIZE 114 50 85mm

Price £27.99 + £2.24 VAT P&P free.

HY400

240 Watts into 4Ω

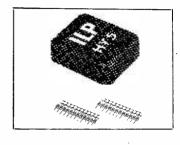
The HY400 is i.L.P.'s "Big Daddy" of the range producing 240W into 4Ω ! It has been designed for high power disco address applications. If the amplifier is to be used at continuous high power levels a cooling fan is recommended. The amplifier includes all the qualities of the rest of the family to lead the market as a true high power hi-fidelity power module. FEATURES: Thermal shutdown-Very low distortion-Load line protection-No external

APPLICATIONS: Public address—Disco—Power slave—Industrial

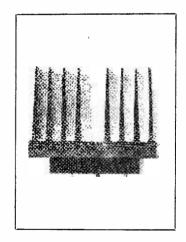
SPECIFICATIONS
OUTPUT POWER 240W RMS into 4Ω LOAD IMPEDANCE 4-16Ω DISTORTION 0·1% at 240W at 1kHz
SIGNAL NOISE RATIO 94dB FREQUENCY RESPONSE 10Hz-45kHz~3dB SUPPLY VOLTAGE ±45V
INPUT SENSITIVITY 500mV SIZE 114 100 85mm

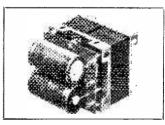
Price £38.61 + £3.09 VAT P&P free.

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Practical WIRELESS

JANUARY 1979 VOLUME 55 NUMBER 1 ISSUE 863

BRITAINS LEADING JOURNAL FOR THE RADIO & ELECTRONIC CONSTRUCTOR

Published by IPC Magazines Ltd., Westover House, West Quay Rd., POOLE, Dorset BH151JG

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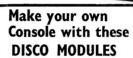




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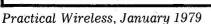
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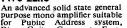
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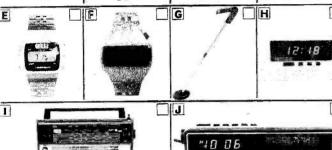
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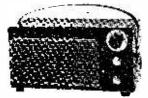


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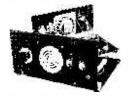
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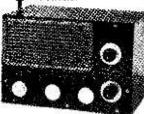


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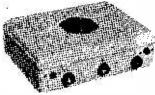
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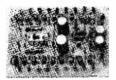
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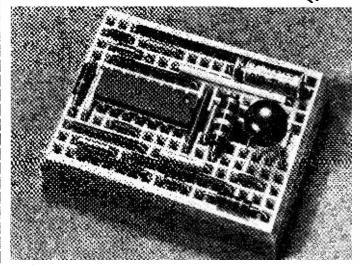
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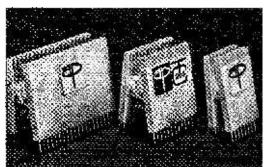
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INTEGRATED CIRCUITS	7460 0 18 7.7470 0 35 7.7472 0 36 7.7472 0 36 7.7473 0 36 7.7474 0 40 7.475 0 54 7.7476 0 40 7.480 0 55 7.7482 0 75 7.7483 0 90 7.7484 1 90 7.7486 0 35 7.7486 0 3	7411 0.80 7411 7492 0.60 7411 7493 0.60 7412 7494 0.60 7412 7495 0.72 7412 7496 0.80 7412 7497 3.00 7412 7417 7417 7417 7410 0.50 7412 7410 0.50 7414 7411 0.50 7414 7411 0.50 7414 7411 0.50 7414 7411 1.75 7414	9 1.50 7 0.83 7 20 0.80 7 21 0.40 7 21 0.80 7 33 1.80 7 55 0.55 7 7 22 0.70 7 88 0.60 7 7 10 0.80 7 7 11 0.80 7 7 12 0.80 7 7 7 12 0.80 7 7 12 0.80 7 12 0.80 7 7 7 7 8 8 0.80 7 7 8 8 8 0.80 7 7 8 8 8 0.80 7 8 8 0.80 7 8 8 0.80 7 8 8 0.80 7 8 0.80	4155 0-85 4156 0-85 4157 0-75 4159 2-10 4170 2-30	74179 1 · 25 74160 1 · 15 74190 1 · 50 74191 1 · 50 74192 1 · 35 74193 1 · 35 74194 1 · 25	74196 1 20 74197 1 10 74198 2 25 74199 2 30° 74199 2 30° 7419	TBA540Q 2-30° TBA550Q 3-22° TBA673 2-19° TBA700 1-52° TBA720Q 2-30° TBA820Q 2-00° TBA820Q 2-90° TBA820Q 2-90° TBA820Q 2-90° TBA820Q 2-90° TBA920 2-90° TBA920	TBA990Q TCA270Q 2 - 99* TCA760A 1 - 38* DIL Sockets 8 PIN 0 - 15 14 PIN 0 - 17
Valve screening 3KP1* 15-00 VCR	1* 10.00 V0 1A 40.00 T1 15A 5.00 *= 7* 5.00 *= 7-5 25.005 36.00 11 68.08	CR517B* 6.00 CR517C* 6.00 ube Bases 0.75 Surplus AT 8%	A49-191/19 A51-110X A56-120X A55-14X	2X BLE FROM	42 · 32 44 · 00 45 · 72	A63-200X A66-120X A67-120X A67-150X	ULLY GUA	52·50 53·61 53·61 54·60
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AL120

AMPLIFIER (With integral short-circuit





OUTPUT POWER	50 Watts R.M.S.
SUPPLY	70 Watts
LOAD IMPEDANCE	8-16 ohms
TOTAL HARMONIC DISTORTION	05% Max. (Typically 02%)
FREQUENCY RESPONSE 148	25Hz-20kHz
SENSITIVITY	500mV
MAX HEAT SINK TEMP.	45 deg. C
DIMENSIONS	192 x 89 x 49 mm

+ 8% V.A.T. P. & P. 35p Introduced to fulfill the demand for a fully protected power amp., capable of driving high quality speaker systems at up to 50w. with distortion levels below 05%. Ideal for domestic use. Discos, P.A. systems, electronic organs etc. The generously rated components ensure continuous operation at high output levels.

SPM120 STABILISED POWER SUPPLIES

SPM120/45 SMP120/65





AC INPUTS	
SPM120/45	40-48v
SPM120/55	50-55v
SPM120/65	60-65v
OUTPUT CURRENT	2.5A
RIPPLE	1A 100mV 2A 150mV

SPM120 is a fixed voltage stabiliser available with an output voltage of either 45v, 55v, or 65v. Designed primarily for use in audio applications, the stabiliser which provides output currents up to 2.5A., operates direct from a mains transformer requiring only the addition of 2 Electrolytic capacitors to complete the s/c protection.

GE100 Mk2. 18 CHANNEL MONOGRAPHIC

f20.00

EQUALISER



Control Range	<u>+</u> 12d8
Dynamic Range	110dB
Maximum Output	+15dB
Frequency Response	30Hz-20KHz (±1dB)
Power Supply	15-0-15v.
Voltage Handling Input	3v R.M.S.
T.H.D.	005%

Only 155mm x 65mm including the 10 x 10K lin slider potentiometers and knobs which are mounted on a board positioned above the circuitry. In the frequency range of 31Hz to 20KHz you can cut and boost ±12dB with the 10 sliders, each of which has its frequency marked on the circuit board. The GE100 has numerous uses including mixers, P.A. systems and discos. It will also greatly improve the sound reproduction of your existing audio equipment. Power Supply for GE100, o/d SG30 £3.80.

VPS30 REGULATED VARIABLE STABILISED POWER SUPPLY +8% V.A.T.



AC Input Maximum	25v	
Voltage Regulation	2-30v	
Regulated Current	0-2A	
Incorporating short circuit pr	otection	

20Hz to 20kHz x 1dB

100mV/100 K ohms

3-5mV/50 K ohms

Within ± 1dB from 20Hz to 20kHz

+ 15dBs at 75Hz

35 to 706v

+ 10-20dBs at 15kHz

Better than 65dBs (All inputs)

Better than 2dBs (All inputs)

300 x 90 x 33mm (less controls

Less than 1% (Typically -70%) 100mV/100 K ohms

500mV

This NEW versatile Regulated Variable Stabilised Power Supply with short circuit protection and current limiting, is a must for all electronics enthusiasts. It incorporates adjustable voltage from 2v-30v, with a current limiting range of 0-2A. With this module there is no need to build a separate power supply for each of your projects, with the simple addition of a transformer (o/d 2033), 0-1ma (o/d 1310 or 1305), plus a suitable shunt, a voltmeter (o/d 1311 or 1306), a 470ohm pot (o/d 1896), a 4K7 pot (o/d 1899), it can be used again and again as a self-contained bench, power supply, eliminating the use of batteries and thus saving

SENSITIVITY

FOUALISATION

FREQUENCY RESPONSE

RASS CONTROL BANGE

SIGNAL/NOISE BATIO

INPUT OVERLOAD

TREBLE CONTROL BANGE

TOTAL HARMONIC DISTORTION

1. TAPE 2. RADIO TUNER

PA200

STEREO PRE-AMPLIFIER



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£16.55

DIMENSIONS The PA200 is basically our popular PA100. Modifications have been made to make it compatible with the higher output AL120 and

SUPPLY

HEADPHONES

A top quality headphone with cushioned earpads and headband. Separate balance/volume controls. Stereo or Mono switch. Impedance: 8 ohms. Frequency; 30-18,000Hz. o/n 884. £8.70. + 12½% V.A.T. p&p 70p.

TO, OUGHZ, O/n 884, £8.70. + 12½% V.A.T. p&p 70p. A brilliant compromise between price and performance. Superb stereo reproduction for the newcomer to Hi-Fi. Impedance 8 ohms. Frequency: 30-15,000Hz. o/n 885. £4.40. + 12½% V.A.T. p&p 50p.

BIB HI-FI ACCESSORIES

Parallel Tracking GROOV KLEEN
The very latest in automatic record cleaning. Designed to suit all modern single play decks. Simple to fit it is extremely efficient. Complete with two types of base and three height extensions. o/n 8101. £3.68.
+ 8% V.A.T. p&p 35p.

Cassette Tape Editing Kit
Enables cassette tapes to be edited and joined easily,
quickly and accurately. Kit comprises: Tape Splicer §
(3.2 mm). 2 Precision Tape Cutters. Tape Placen.
(3.2 mm). 2 Precision Tape Cutters. Tape Placen.
Self-adhesive Labels. Real of Splicing tape. 3
Winders and removers and instructions, all in a
handy wallet. o/n 811. £2.40. + V.A.T. p8p 35p.

GROOV-STAT
The BIB Groov-Stat static reducer neutralises the static charge on records and other plastic surfaces. o/n 8103, £5.45. + 8% V.A.T. p&p 35p.

Cassette Head Cleaner Essential for cleaning of tape heads, capstans and rollers. Pack contains 1 pape Head Applicator and tape head polisher tools. Plus bottle of special formula cleaning fluid and full instructions. o/n 832, £0.56. + 12½% V.A.T. p.8 p.35p.

ADAPTORS

AC-DC enables a large range of battery powered radios, recorders, calculators to be run off the mains, (220-240v AC) Switchable for 6, 7.5 or 9 volts, Current rating 2,500mA, Polarity reversing switch. Universal plug incorporated. o/n 137. £3.95. + 12 ½% V.A.T. p&p 35p. DC-DC for use in all cars, boats etc., with pos. or neg. earth for a regulated output of 6, 2.5 or 9 volts DC at 1A max. For radios. recorders etc. o/n 138. £2.80. + 12}% V.A.T., p&p. 32p.

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2-WAY channels for high and low frequencies to correct speakers – high to tweeters, low to woofers. Complete with instructions. Frequency: 3,000Hz, o/n 1904. £1.10. + $12\frac{1}{2}\%$ V.A.T, p&p 35p.

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CASES

TEAK 30, 32 \times 23 \times 8cm, designed mainly for use with our stereo 30 Audio System but has proved very helpful to home constructors. Fitted with solid uncut front and back. o/n 139, £5.45. + 12 $\frac{1}{2}$ % V.A.T. p&p 70p.

TEAK 60, 42 \times 29 \times 9cm, for use with AL60/MK60 Audio Kit. Useful for the home constructor requiring an amplifier sleeve — has no front or back panel. o/n 140. £7.00. + $12\frac{1}{2}$ % V.A.T. p&p 85p.

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Miniature Balance & Tuning Meter
Miniature moving-coil meter for stereo balance
indicator, tuning indicator for FM or similar
application. Pointer at centre indicates zero or null
position. Robust construction. Sensitivity;
100-0-100M. Dimensions: 23 × 22 × 26mm.
o/n 1318. £1.95 +8% V.A.T. p&p 35p.



Balance and Tuning Meter Clear view edgewise meter. Centre zero application. Sensitivity: 100–0–100UA Dimensions: 45 x 22 x 34mm. o/r 1319. £2.00. +8% V.A.T. p&p 35p.

Miniature Level Meter
Moving coll, for accurate level indication for tape recorders, amplifiers etc. Neat design, rugged construction – will withstand five times rated value. Sensitivity: FSD: 200UA. OdB: 130UA. Dimensions: 23 x 22 x 26mm. o/n 1320. £2.80. +8% V.A.T. p&p 35p.



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DYNAMIC CASSETTE

DYNAMIC CASSETTE
For equipment requiring a high quality microphone. Sturdy, solid
moulded body in black with neat chrome surround. Pick-up pattern is
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and 3.5mm plugs. Matching moulded strut. Impedance: 200 ohms.
Sensitivity: 90aB. Frequency; 90-10.000Hz. Size: 20mm dia x
120mm.o/n1326.21.50. + 12 \(\frac{1}{2} \)% V.A.T. p&p 35p.

DYNAMIC MICROPHONE
Superior quality portable cassette recorder mike with built-in remote control switch and lead fitted with 5-pin 240° DIN plug fremote switch) and 3-pin DIN plug Imicrophonel. Provides a direct replacement for those supplied with recorders. With detachable stand. Omnidirectional. Impedance: 200 ohms. Freq. response: 100 to 10,000Hz. Sensitivity: 79dB at 1,000Hz. o/n 1327. £2.65. + 12½% V.A.T. p&p 35p.

RE - 317: DYNAMIC MICROPHONE

RE – 317: DYNAMIC MICROPHONE Highly sensitive, high-grade desk or hand mike suitable for use with many popular cassette decks, incorporates Or/Off switch and 1 metre lead with moulded standard jack plug. Complate with desk stand. Omnidirectional. Impedance: 5,000 ohms. Freq. response: 100 at 12,000Hz. Sensitivity: (-7dB at 1,000Hz). o/n 1,336. £4.10. + 12½% V.A.T. p&p 35p.

OMNIDIRECTIONAL CARDIOID
Powered by a 1½v battery located within the aluminium body. Satin silver finish with front disk protection to the diaphragm housing. On/Off switch. Also with "Busby type windshield," "U" bracket and stem and extremely supple cable. Consumption: 0.2mA from 1½v battery providing approx., 8-10,000 hours continuous lite. Impedance: 600 ohms. Sensitivity: 70dB. Frequency: 30-16,000Hz. Size: 23mm dia + 267mm. o/n 1329, £12,80. 12½% V.A.T. p&p 35p.

UNIDIRECTIONAL CARDIOID
Dual imp. 600 and 50,000 ohms. Response 50 to 14,000Hz.
Sensitivity 54dB at 50K/ohms. Size: 1\frac{1}{4}" dia \times 6\frac{3}{4}" long. Weight approx., 190gm. o/n 1328. £10.95. + 12\frac{1}{4}% V.A.T. p\frac{8}{4}p 35p.

STANDS

GOOSENECK CHROME FLEXIBLE HOLDERS Length 320mm. o/n 1333. £2.40. + 12½% V.A.T. p&p 35p. Length 515mm. o/n 1334. £3.40. + 12½% V.A.T. p&p 35p.

FLOOR STAND Heavy chrome. Stow-away feet with rubber ends for maximum stability. Draws to a height of 5' maximum, o/n 1335. + 12½% V.A.T. p&p 85p.

BOOM ARM for use with the above stand. Heavy chromed metal, it gives 30" reach from the stand. o/n 1337 £8.00. + $12\frac{1}{2}\%$ V.A.T. p&p 70p.

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123	with inline fused power lead and instructions. 6.6m Coiled Guitar Lead Mono Jack Plug	£0.60*
124 125	to Mono Jack Plug BLACK 3 pin DIN plug to 3 pin DIN plug. Length 1.5m 5 pin DIN plug to 5 pin DIN plug. Length 1.5m	£0.75° £0.75°
126 127	5 pin DIN plug to Tinned open end. Length 1.5m 5 pin DIN plug to 4 Phono Plugs.	£0.75* £1.30*
128 129	Alf colour coded. Length 1.5m 5 pin DIN plug to 5 pin DIN socket. Length 1.5m 5 pin DIN plug to 5 pin DIN plug mirror	£0.80*
130	image, Length 1.5m 2 pin DIN plug to 2 pin DIN inline socket, Length 5m	£1.05* £0.68*
131	5 pin DIN plug to 3 pin DIN plug 1&4 and 3&5. Length 1.5m 2 pin DIN plug to 2 pin DIN socket. Length 10m	£0.83*
133	5 pin DIN plug to 2 phono plugs. Connected pins 3&5. Length 1.5m	£0.75°
134	5 pin DIN plug to 2 phono sockets. Connected pins 3&5, Length 23cm	£0.68*
135	5 pin DIN socket to 2 phono plugs. Connected pins 385. Length 23cm	£0.68*
136	Coiled stereo headphone extension lead. Black, Length 6m AC mains lead for calculators etc.	£1.75* £0.45*
178	Please add 8% V.A.T. to all the above.	20.40

High quality audio modules for Stereo and mono

S450

STEREO FM TUNER Fitted with phase lock-loop





FREQUENCY RANGE	88-108 Mhz
SENSITIVITY	3 0 µV
BANDWIDTH	250 kHz
SPURIOUS REJECTION	50 dB.
SELECTIVITY ± 400 kHz	55 dB
AUDIO OUTPUT (22 5 kHz devi	ation) 100 mV
STEREO SEPARATION	30 dB
SUPPLY REQUIREMENTS	20 to 30V (90mA max)
AERIAL IMPEDANCE	75 ohms
DIMENSIONS	240mm × 110mm × 32mm

The 450 Tuner provides instant programme selection at the touch of a button ensuring accurate tuning of 4 pre-selected stations, any of which may be aftered as often as you choose, simply by changing the settings of the pre-set controls. Features include FET input stage. Vari-Cap diode tuning. Switched AFC LED Stereo Indicator.



OUTPUT POWER	7 Watts RMS
LOAD IMPEDANCE	8 ohms
TOTAL HARMONIC DISTORTION	Less than 5% (Typically 3%)
FREQUENCY RESPONSE	50 Hz to 20 kHz ± 3dBs
TONE CONTROL RANGE	± 12 dBs at 100Hz and 10kHz
SENSITIVITY	190 mV for full output
INPUT IMPEDANCE	1 M ohms
TRANSFORMER REQUIREMENTS	22 V.A.C. rated at 1A
(Less controls and panel)	200mm × 130mm × 33mm

The Stereo 30 comprises a complete stereo pre-amplifier, power amplifiers and power supply. This, with only the addition of a transformer or overwind will produce a high quality audio unit suitable for use with a wide range of inputs i.e. high quality ceramic pick-up, stereo tuner, stereo tape deck etc. Simple to install, capable of producing really first class results, this unit is supplied with full instructions, black front panel, knobs, main switch, fuse and fuse holder and universal mounting brackets.



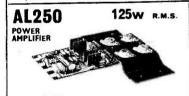
OUTPUT POWER	25 Watts RMS
SUPPLY	30-50 V
LOAD IMPEDANCE	8–16 ohms
TOTAL HARMONIC DISTORTION	Less than 1% (Typically 06%
FREQUENCY RESPONSE	20 Hz to 30 kHz × 2 dBs
SENSITIVITY	280 mV for full output
MAX. HEAT SINK TEMPERATURE	90°C
DIMENSIONS	103mm × 64mm × 15mm

This high quality audio amplifler module is for use in audio equipment and stereo ampliflers and provides output powers up to 25 RMS with distortion levels below 0.1%.



OUTPUT POWER	35 Watts RMS
SUPPLY	40-60 V
LOAD IMPEDANCE	8-16 ohms
TOTAL HARMONIC DISTORTION	Less than 1% (Typically 06%
FREQUENCY RESPONSE	20 Hz to 30 kHz × 2 dBs
SENSITIVITY	280 mV for full output
MAX. HEAT SINK TEMPERATURE	90°C
DIMENSIONS	103mm × 64mm × 15mm

The AL80 is similar in design to the AL60 above and is of the same high quality but provides output powers up to 35W with distortion levels below 0.1%.



125 Watts RMS continuou
50-80 V
4-16 ahms
25 Hz 20 kHz measured at 100 Watts
450 mV
33 K ohms
0.1%

£17.82* + 40p p&p + 8% VAT This unit, designated AL250, is a power amplifier providing an output of up to 125W RMS, into a 4 ohm load.

AL30A AUDIO AMPLIFIER MODULES

£3.79



MAXIMUM SUPPLY VOLTAGE	30 V
POWER OUTPUT for 2% THD	10 Watts RMS
TOTAL HARMONIC DISTORTION	Less than ·25%
LOAD IMPEDANCE	8-16 Ohms
INPUT IMPEDANCE	100 K ohms
FREQUENCY RESPONSE	50 Hz-25 kHz ± 3 dBs
SENSITIVITY	75 mV for full output
DIMENSIONS	74mm × 63mm × 28mm

These low cost 5 and 10 watt modules offer the utmost in reliability and performance, whilst being compact in size

SPM80 STABILISED POWER SUPPLY £4.40 + 35p p&p + 12:1% VAT



INPUT A.C. VOLTAGE	33-40V
OUTPUT D.C. VOLTAGE	33 V nominal
OUTPUT CURRENT	10 mA-1 · 5 amps
OVERLOAD CURRENT	1.7 amps approx.
DIMENSIONS	105mm × 63mm × 30mm

FREQUENCY RESPONSE 20 Hz to 20 kHz × 1 dB
TOTAL HARMONIC DISTORTION Less than ·1% (Typically ·07%)

100 mV/100 K ohms For an 100 mV/100 K ohms output 3.5 mV/50 K ohms 250 mV

Better than 26 dBs (All inputs)

Within ± 1 dB from 20 Hz to 20 kHz ± 15 dBs at 75 Hz

+ 10-20 dBs at 15 kHz Better than 65 dBs (All inputs)

Designed to power two AL60s at 15 Watts per channel simultaneously. Circuit Techniques include full short circuit protection.



PRE-AMPLIFIER	THE STATE OF THE S
£16.05	

£16.05
+ 40p p&p +12½% VAT

	SUPPLY	2010 40 4
p p&p }% VAT	DIMENSIONS	300 x 90 x 33mm (less controls)
WAT p quality stereo pre-amplifier and tone control frements of stereo amplifiers or audio units. The filters for high and low frequencies.	unit, the PA100 provides a e six push button selector sy	a comprehensive solution to the front end witch gives a choice of inputs together with

FOULLISATION

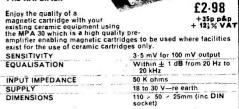
SENSITIVITY 1. TAPE INPUTS 2. RADIO TUNER 3. MAGNETIC P.U.

TREBLE CONTROL RANGE

SIGNAL/NOISE RATIO INPUT OVERLOAD

MPA30

MAGNETIC CARTRIDGE PRE-AMPLIFIER



PA12

STEREO PRE-AMPLIFIER



£7.78 + 30p p&p + 121% VAT

The PA12 Stereo Pre-Amplifier chassis is designed and recommended for use with the AL 20/30 Audio Amplifier Modules, the PS12 power supply and the T538 Transformer, Features include on/off volume, Balance, Bass Table Seaticle Complete with tape output.

and Treble controls. Complete with	tape output.
FREQUENCY RESPONSE	20 Hz-20 kHz (-3dB)
BASS CONTROL	± 12 dB at 60 Hz
TREBLE CONTROL	± 14 dB at 10 kHz
INPUT IMPEDANCE	1 Meg. ohm
INPUT SENSITIVITY	300 mV
CROSSTALK	-60 dB
SIGNAL/NOISE RATIO	65 dB
OVERLOAD FACTOR	± 20 dB
TAPE OUTPUT IMPEDANCE	25 K ohms
DIMENSIONS	152mm × 84mm × 25mm

PS12 POWER SUPPLY MODULE

Power supply for AL20A-30A, PA12, S450 etc. Transformer T538.

Input A.C. Voltage 15-20V Output D.C. Voltage 22-30V approx. (Dependent upon Output Current 800mA

Dimensions 60 × 43 × 26mm.

£1.50 + 12 1/2 VAT + 35p p&p.



BP124 SIREN ALARM MODULE

American Police screamer powered from any 12 volt supply into 4 or 8 ohm speaker. Ideal for car burglar alarm, freezer break-down, and other security purposes.

ONLY £3.50 + 8% VAT + 35p p&p.

MA60 HI-FI AMPLIFIER KIT

Build you own top quality amplifier, save yourself pounds. The MA60 kit comprises the following Bl-kits modules, $2 \times AL60$ amps, $1 \times PA100$ pre-amp, $1 \times SPM80$ stab. power supply. $1 \times BM780$ transt, giving 17 watts RMS per channel STEREO. All modules covered by the Bl-PAK satisfaction or money back guarantee. Details of the above modules are in this ad. Price £32·00 + 12;% VAT + 52p p&p.

TC60 KIT

A beautifully designed genuine TEAK WOOD veneered cabinet to put the professional touches to your home built amplifler. Full set of parts incl. Front & Back Panels, Knobs, Chassis, Fuses, Sockets, Noen, etc. Ideal for the MA60. Size: 425mm > 290mm > 95mm/2011. 95mm. Price £19-95 + 12½% VAT + 86p p&p

TRANSFORMERS

T538 For use with S.450 AL30 A MPA30 Order No. 2036 Price: £3·20 + 55p p&p + 12½% VAT 7250 For use with Stereo 30 Order No. 2050 Price: £3·25 + 55p p&p + 12½% VAT 8MT20 For use with AL50 SPM50 Order No. 2034 Price: £5·40 + 86p p&p + 12½% VAT 8MT250 For use with AL550 Point Stereo SPM50 Order No. 2035 Price: £6·35 + £1·10 p&p + 12½% VAT 9Tder No. 2035 Price: £6·35 + £1·10 p&p + 12½% VAT



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A Case for Change

T is more than seven years since an article describing the construction of a metal detector first appeared in *Practical Wireless*. Since then, as technology has progressed, new types of detector have been developed and appeared on the market, and several more designs have been published in our pages. This month, we present a Pulse Induction detector, the PW "Sandbanks", plus a feature on metal detectors and their use.

It can be argued that metal detecting is not a subject appropriate to *PW*, though it has an affinity to radio or more closely to radiolocation, in that a field is radiated, and the effect of an object within that field is monitored by some sort of receiving system. One thing which cannot be disputed is the amazing rise in popularity of this hobby of searching for buried objects. During 1977, the Home Office issued some 24,000 "Pipe Finder" licences. In 1978, the figure had reached 30,000 by September, and during October licence applications were flooding in at a rate twice that of the 1977 average. This resulted in the accumulation of a back log of several thousand applications, and a wait of some weeks before licences were despatched.

The suggestion has been put forward that metal detector users could be granted "blanket" licences, authorising them to use any machine which has received Home Office approval. At present, anyone buying or building a new detector has to obtain a new licence, and this must obviously increase the number of applications being dealt with, perhaps unnecessarily. After all, the current system of issuing a licence only for a named brand does no more to prevent the unscrupulous from using a detector which has not been approved.

I wonder how many readers have noticed that *Practical Wireless* issues are numbered in annual volumes which run from May to April? Or at least they did; commencing with this issue, they change to January to December. The previous rather odd arrangement came about when the magazine changed from its original weekly format to monthly in the 1940s. Volume 54 therefore comprises eight issues from May to December 1978, and Volume 55 begins this month. I hope this change will have removed a source of occasional confusion.

May I take this opportunity to wish all our readers a very happy Christmas and a healthy and prosperous 1979, on behalf of all the staff on *PW*.

PERSONALITY



Dick Ganderton—Assistant Editor

Dick joined *PW* a year ago after six years as Features Editor of *Electronics Weekly*. His entire working life has been spent in professional electronics including ten years in the guided weapons business. He turned to technical journalism, still in electronics, some twelve years ago.

Much of his spare time is devoted to model engineering and serious railway modelling to P4 and Scale Seven standards. As an active member of the EM Gauge Society he edits their journal. Other interests are radio control, dinghy sailing, and a Honda 400/4.

Married for fifteen years to a wife who mostly tolerates his long hours in 'the shed', he has a daughter and two sons. Dick now lives in Verwood, enjoying the peaceful life after the hectic existence of working in London, and missing the privilege of helping to keep British Rail insolvent.

NEWS...

NEWS...

NEWS...

'Sounds Vintage'

Readers who are interested in early electronic equipment will be pleased to know a magazine is to be launched exclusively for the vintage enthusiast.

'Sounds Vintage' will be published bi-monthly, catering for those interested in vintage wireless, gramophones and cylinder machines, vintage amplifiers, pre-war literature, practical hints on the care, maintenance and restoration of vintage equipment, news from the major auction rooms, readers' letters and wants, and—in fact—anything which will be of interest to the collector.

Starting off as a 32-page A4 presentation, 'Sounds Vintage' will be available on subscription only. Issue No. 1 is scheduled for publication in mid-January 1979. The annual subscription will be £5.80 inland, £6.80 overseas, postage paid. A special offer is being made for issue No. 1 only—a sample single copy at 65p post-paid from: 'Sounds Vintage', 28 Chestwood Close, Billericay, Essex.

Technical Literature

General Instrument Microelectronics have issued a new product guide describing the company's complete range of MOS-LSI microcircuit products.

The 24-page guide lists product types, functions, brief parameters and special features in many different application areas, including microprocessors, t.v. games, domestic appliances, industrial, radio & t.v., telecommunications, etc. It also includes details of GIM's MOSFET range.

Copies may be obtained free of charge from G.I.M. Ltd., Regency House, 1-4 Warwick Street, London W1R 5WB. Tel: 01-439 1891.

The new guide to Mullard 'preferred' power transistors is now available. Categories included are complementary power transistors, complementary Darlingtons, *npn* high-voltage transistors, *npn* switching power transistors and transistors for switched-mode power supplies. Also included in the guide are recommended power transistor line-ups and an equivalents list.

Requests for copies should be addressed to Central Enquiry Handling Unit, Mullard Ltd., New Road, Mitcham, Surrey.

CRT investment

Mullard Ltd. have announced details of a £24m investment programme to further improve manufacturing facilities at the company's colour t.v. tube assembly plant at Durham and to establish a 20inch 90° tube production line at their Simonstone plant, to be known as Project Vanguard.

Backed by a £4.5m grant under the Government's industry support scheme, the programme covers a three-year period. Investment at Simonstone will total £13.1m and at Durham £7.8m.

Further investments of £2·4m will be made at Washington, which produces the neck components for the tube, and £0·9m at Crossens where the related magnetic components are manufactured. Together, these plants employ some 4000 people engaged directly or indirectly in t.v. tube production.

Mullard is now the UK's only manufacturer of picture tubes, following the closure of the Thorn plant in 1977.

New catalogues

Carel Components latest Product Guide and Price List is now ready for distribution.

The catalogue, which covers Carel Components' comprehensive product range of 'Sure Connections and Practical Devices' is obtainable free upon request from: Carel Components, 40-44 The Broadway, Wimbledon SW19 1SQ. Tel: 01-540 7186.

Home Radios' latest catalogue is now available. A number of items for which the company experienced little demand have been dropped, resulting in the catalogue being a little slimmer. However, the company hope to compensate for this by expanding the 'Bargain List' over the next few months.

The price is £1.00 plus 25p P&P from: Home Radio (Components) Ltd., 234-240 London Road, Mitcham, Surrey CR4 3HD, Tel: 01-648 8422.

Ace Mailtronix Ltd. mail order catalogue issue 2, covers components kits, modules, tools, test equipment and audio accessories. The catalogue costs 30p, which is refundable with an order of £5.00 or over.

Following the policy change by Doram, not to supply R.S. Components products (see News column — November 1978), Ace Mail-

tronix Ltd. have undertaken to obtain any current R.S. item subject to a minimum order of £2.00.

For details of prices of R.S. items and applications for the catalogue, please contact: Ace Mailtronix Ltd., Tootal Street, Wakefield, West Yorkshire WF1 5JR. Tel: (0924) 250375.

Verospeed the fast-turnround component suppliers, inform us that their latest catalogue is now available. To obtain this free, expanded catalogue, apply to: Verospeed, Barton Park Industrial Estate, Eastleigh, Hants. Tel: (0703) 618525.

The latest Radio and Electronics Books catalogue is now available from Bernard Babani (publishing) Ltd., if you apply enclosing a stamped addressed envelope. Bernard Babani (Publishing) Ltd., The Grampians, Shepherds Bush Road, London W6 7NF. Tel: 01-603 2581.

Vero push retail side

A new, bright packaging style has been adopted by Vero Electronics to protect and promote the range of their products offered to the home constructor. The larger boxes will be individually packed in cartons, while smaller items will be bubble packed. The extra costs of bubble packing will be offset by reduced handling costs during manufacture, so that the end price to the constructor will not alter.

A catalogue of Vero Products for the Home Constructor is available, price 10p, from Vero stockists or direct from: The Retail Dept., Vero Electronics Ltd., Industrial Estate, Chandler's Ford, Hants SO5 3ZR.





This part of the Dorchester project deals with the construction and the complete circuit is shown in Fig. 7. To those of you weaned on valves—and even transistors—the TDA1090 should be something of a revelation in terms of parts-count shrinkage. In keeping with the policy of designing a readily repeatable unit, all switching is performed with devices which fit directly on to the p.c.b., thus greatly reducing the chance of mis-wiring such a complex arrangement.

The fact that the TDA1090 requires only single coupling to the oscillator coil is a significant factor in making this design possible. As it is, some underboard link wiring is essential if some very necessary earth plane areas are not to be chopped and become parasitic v.h.f. tuned 'lines'.

The coils are selected from the recently expanded range of Toko short-wave devices and fit between the switch modules. Their impedances are generally somewhat higher than those usually associated with bipolar circuitry, since the input resistance of the a.m. mixer stage is relatively high at $5k\Omega$.

Each coil is provided with a trimmer, with the exception of the l.w. coil on the rod antenna. A ferrite rod is used for the usual m.w./l.w. as it offers better immunity to the electrostatic interference associated with wire antennas at these frequencies. A coupling coil may nevertheless be provided if required, but a separate antenna to that used for short waves is advised.

The i.f. filtering for the a.m. system employs a Toko MFH series mechanical unit. These are available in many bandwidths and the board is adaptable to the CFU and CFM 2 ceramic and mechanical series.

In this project, a 4 or 5kHz bandwidth is used. Whilst this may seem low for the usual m.w. and l.w.

channel spacings, it is quite necessary for s.w. reception, where all the channels are far closer and likely to cause mutual interference.

The only additional points of note with regard to f.m. are the quadrature feed choke and low pass (birdie) filter. The quadrature choke is a tuneable screened type L132, as opposed to the more usual fixed variety. It performs a dual function—primarily screening the radiation of 10·7MHz squarewaves from the rest of the circuit, but also providing a means of adjustment to the detector injection level on f.m., thus achieving optimum muting operation.

The birdie filter used is of LC construction. Too many modern designs overlook the basic benefits of the LC filter in this position, favouring instead more complex (and often more ineffective) active arrangements. The pi-LC filter provides a virtually flat response from d.c. to 55kHz and, most important, it effectively decouples all the remaining i.f. and i.f. harmonics from the audio which is fed to the decoder stage. A pre-decoder pre-amplifier drives the 1310 circuit with a reasonably high level of composite signal to ensure a good signal-to-noise ratio. Output terminations along one edge provide the majority of connection points.

Construction

The component layout (Fig. 8b) describes the location of all parts used in the tuner. A printed circuit of the given design is virtually essential if repeatable performance is to be achieved, the layout of r.f. stages being particularly critical.

Fig. 7: (right) The complete circuit diagram

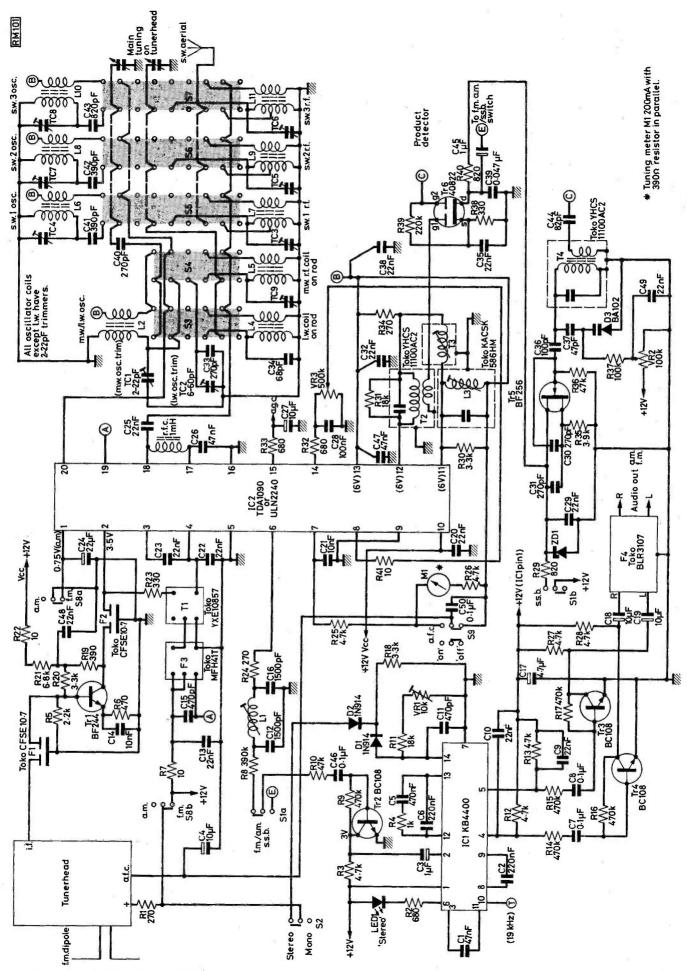
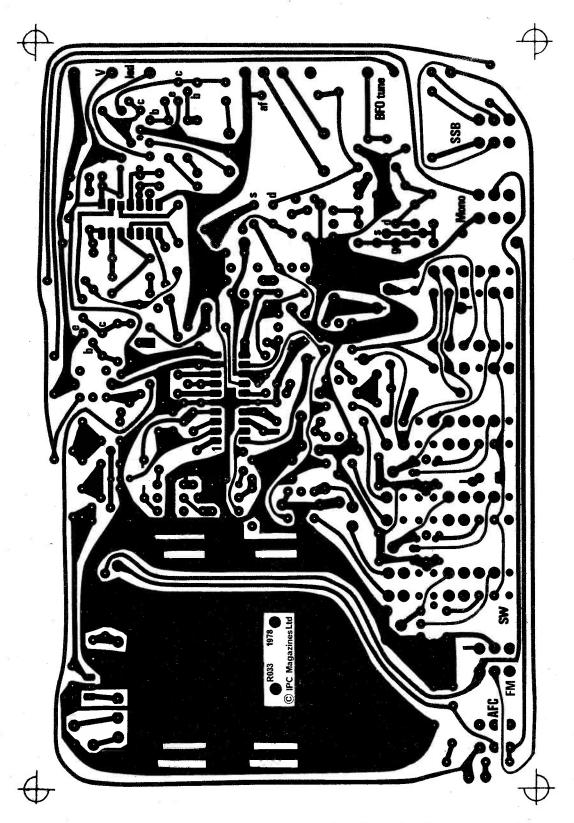
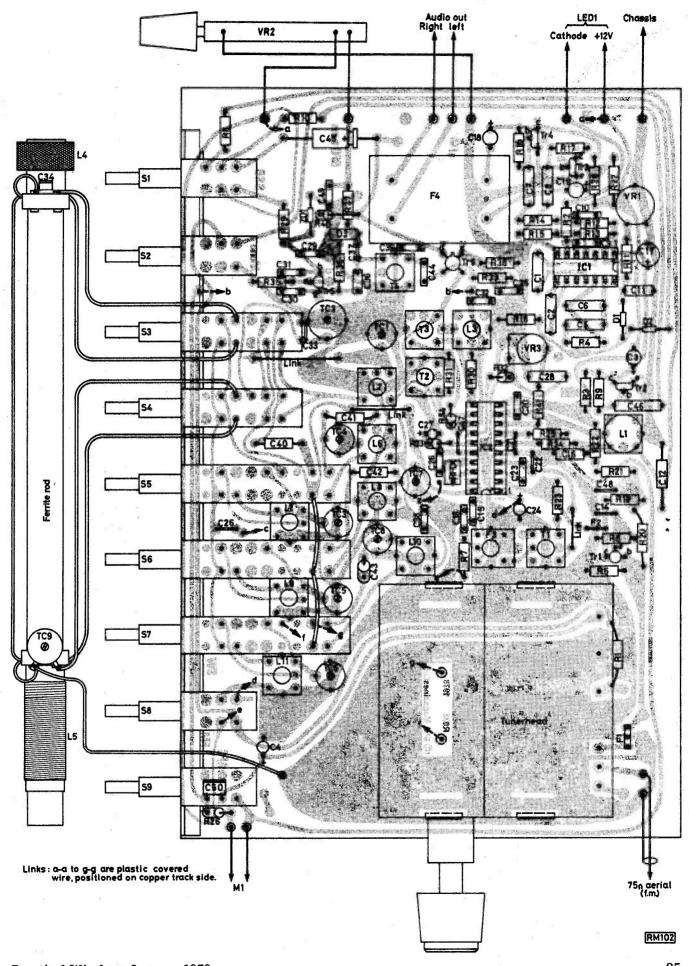


Fig. 8a: (below) The copper side layout of the p.c.b. Fig. 8b: (right) The p.c.b. component placement layout



As ever, the prime 'caveat' for the constructor is to ensure good soldering. This may sound repetitive to those of you who have heard it all before, but soldering is the first and last word in any electronics project. Far too many theoreticians in electronics are truly hopeless when it comes to wielding a soldering iron.

Always use an iron that is hot enough. Broadly, this means one which doesn't leave great blobs of flux at the joints, since this should evaporate in the



★ components

Resistors			Semiconduct	ors _{No}		
₹ ₩ 10% me	lal oxide 🔭	A STATE OF THE STA	Transistors			es kana
10Ω 270Ω	3 3 *	R7,92,41 R1,24,34	BC108	3	Tr2, 3, 4	10 mg (10 mg) 10 mg (10 mg)
330Ω	2	R23, 38	* BF224 BF256	1	Tri Tri5 - ≠	
390Ω	1 9	' R19	40822	1	Tr6	
470Ω 680Ω	1	R6- R2, 32, 33 *		h		40 79 70
820Ω	2	R40, 29	Integrated cin KB4400	:ujis, * 1	IC1	
.1k0 *,_ :	"1	R4 * *	TDA1090		. Design	e garage
2·2kΩ 3·3kΩ	1000	R5 R18, 20, 30	(or ULN2240)	1	IC2	475T
3-9kΩ ·	-i	- R35	Di odes .	В.,	N	
4·7kΩ 6·8kΩ	. 7	R3, 12, 13, 25, 26, 27, 28	-1N914	2 2	D1/2	5.4
18kΩ	2 .	R21 R11, 31	BA192 (cr.BB105)	4 ** >	D8	
47kΩ	2	R10736 *	T1L209	1	LED1" .*	1.5
100kΩ 220kΩ	1	R37 R39	Zener Diodes			
390kΩ	.1	R8	zeņer Diodes 7V5	1	ZD1	100
470kΩ#	. 5	R9, 14, 15, 16, 17	(BZY88-serie	s).		22
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apacitors			Transformers	``		
	or low-K Ce	ramic. Minimum 30V wkg.	≥ T1 T2 4	Toko YA	10857 (Ambii 5 11100 ACP) (Ambit) (2 off).
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390pF	2	C41, 42	Colls 1	**Toko CA	N 1980 (Ambit	Ϋ́
470pF 820pF	.4 -4	C11, 45 C43	L2 "	Toke YMI	RS 80046N (A	
1500pF	2	C12,16 3 5	1.3 141	Ambit L1		290
47nFix 220nF	1,	C47		Ambit MY	/C (f.w. anten VC2 (m.w. ant	enna coli)
470nF	1	· (C2, 6 · (C5 ·	a. L6	Ambit SV	/O1	W. Trill
0·1μ F	4	►C7, 8, 46, 50	L72;	Ambit SV		, 1967 60
Sub-miniahi	re ceramic r	nin. 30V wkg. "	⁶ . L9	Ambit SV	/R2	
82pF	2	C43, 44	L10	Ambit SV		100
100pF	1	C36	W J	Ambit SV	rno va	
10nF 22nF	13	C14, 21 C9, 10, 13, 20, 22, 23, 25, 29, 32,	Filters		Na .	
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process, leaving only a small residue. The hotter the iron, the quicker the joint will be made—and the brighter and better your work will become. The temperature should be such that the instrument will just scorch a piece of wood without charring it rapidly.

Before fitting the tunerhead, remember to attach a couple of wires to the connection of the a.m. gang (Fig. 9) to feed through the underside of the board. Connection may be made to the top of the tuning gang, but the additional lengths of wire used in the process will be an unnecessary source of stray capacity and unwanted pickup.

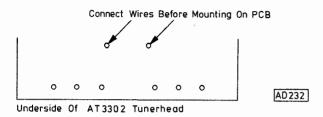


Fig. 9: A.M. tuning gang sections

The a.m. trimmers should be removed from the gang, by unscrewing the points shown in Fig. 10: they are provided for each individual range. The necessary underboard linking, is given in Fig. 8b (don't forget the links on the top of the switch, also shown in the basic layout plan. As a general guide, always fit components with the most leads first, thus narrowing down the numbers of holes for incorrect insertions.

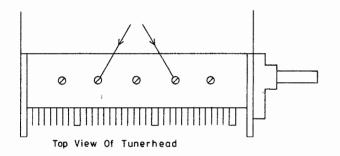


Fig. 10: A.M. trimmers to be removed

Testing and alignment procedures

The frequency ranges for the appropriate bands are listed in Table 1, part 1. The f.m. tuner is fully pre-aligned, and only the i.f. output coil of the tunerhead is likely to require adjustment. The f.m. detector stage may be readily aligned in conjunction with the built-in tuning meter, using the i.f. white noise to centre the detector stage on the centre-zero tuning meter.

The mute pre-set must be fully anti-clockwise during this process, and the core of L132 about 1 to 2 turns below the top of the assembly.

With the f.m. i.f. detector aligned, and using white noise, turn the mute control to its threshold—i.e. until the tuner just mutes. If proper muting cannot be achieved, then a capacitor of between 220 and 330pF should be placed across the output side of the first f.m. ceramic i.f. filter to reduce the i.f. gain. This does not adversely affect the f.m. sensitivity.

Alternatively, L132 could be trimmed, but please note that this approach will require re-tuning of the detector stage quadrature coil after each adjustment. A meter to monitor the control voltage at pin 14 will indicate the status of this part of the circuit, and mute will occur when the voltage at pin 14 is in the region of 1.5 to 4.5 volts.

Since the f.m. tuning meter is centre zero, note also that another amplitude-related function takes place at the a.g.c. terminal, pin 15. However, a.g.c. onset does not occur until a fairly sizeable level of signal is present $(100\mu\text{V}\ \text{or so})$, and so this is not as effective a feature as on the CA3089E.

With a.m., the first points to align are the i.f. stages. The use of a ceramic or mechanical filter arrangement will simplify this process quite considerably since a wide-band signal injector may be used, simply adjusting the a.m. detector coil for maximum noise. If this is not available a finger placed on the input to the a.m. mixer stage will be found more than adequate in all but the most remote of areas.

The mechanical i.f. filter will also need peaking to match the circuit—but not more than a couple of turns. With the a.m. i.f. functional, select the m.w. pushbutton, and tune around. The m.w. antenna coil should be about 3-5mm over the end of the rod, with the trimmer capacitor across this coil about 20% meshed. The oscillator trimmers for all a.m. stages should be initially set to minimum capacitance.

It should not be too difficult to find out the a.m. frequencies of local stations so, with the tuning capacitor out of mesh, turn the core of the m.w. oscillator coil until it is lined up on a known station at the top end of the band. Mesh the trimmer approximately 30%, and re-adjust the core to recover the original station. Turn now to the l.f. end of the m.w. and find a station. This time peak the signal by adjusting the position of the coil on the ferrite rod. Return to the h.f. end, and trim the antenna trimmer. Repeat the process until no further improvement can be obtained.

Next month concludes this article with mechanical details and some useful thought on extending the functions of the receiver.



A REVIEW OF RECENT DEVELOPMENTS

In general, the author does not have any more information on products than appears in the article

Rising solar cells

Japan is often called the Land of the Rising Sun and its current interest in solar cell technology supports this. Among the many newer developments reported is a process for fabricating solar cells but using a low temperature technique which could lead to low cost cells. Instead of the more usual 1,000 C diffusion temperatures commonly required for the conventional diffusion process when making silicon pn cells, the new method needs temperatures of only 300-350 C. A film of tin dioxide, less than 100 Angstroms thick, is deposited by evaporation at 300 C to form a hetrojunction with the n-type silicon. A further deposition at 350 C of a conductive film (about 1,000 Angstroms of it) completes the process. The conversion efficiencies are said to be easily comparable to those of the more conventionally produced silicon cells. With a polycrystalline substrate this is around $8\frac{1}{2}\%$, while single crystal substrates allow this figure to rise to 12%.

Wire edged terrier

If you're ever in the countryside, and you get this strange feeling that you're being "watched", it could be the fence! This latest attempt at a barbed wire eveball is the product of a British company. Its barbed wire fence is different to others in that it has an optical fibre in each wire. If the wire is broken, the system will not only warn of this but can also give the location of the break. A permanent "all quiet on the Western Front" code is transmitted along the fibre, and any interference with our skinny optical friend anywhere along its length (like trying to bridge it) will disturb the code and sound an alarm.

Little repeater stations are located at strategic intervals and these are optically coupled to the fibre. The system puts a special coded signal along the wire; transmitted at one end, received at the other. When a break in the wire occurs, the next little repeater along the line doesn't receive the signal. To register its disappointment it puts out a warning signal of its own, thus locating the break. Because of the complexity of the coded

signal, some 8,000 sections of fencing can be guarded by a single installation. A system to keep an eye on a small garden might cost near to £1,000. Personally, I prefer a Doberman with a grudge and a nervous disposition.

Silly Saph

Diamonds may be a girls best friend, but to computer manufacturers it could be sapphire. Silicon-on-Sapphire (or s.o.s.) is fast gaining recognition and application as the latest central processing unit (c.p.u.) indicates. Previously, the c.p.u. took nine boards and occupied some 700 square inches. The new s.o.s. c.p.u. takes less than one square inch! Small is not only beautiful, it's fantastic.

Grooveless

How many grooves has a gramophone record-just one, it's a single spiral. This old joke will soon be out of date for all time if the Victor Company of Japan has its way. This organisation is working on a grooveless record suitable for playing both video and digital audio. The pick-up is capacitive by means of a tiny conducting strip on the trailing edge of the stylus and the surface of the disc. Special narrow tracks give a servo signal and this is used to keep the stylus in proper tracking. The discs themselves use a conductive plastic and this, of course, can be simply "pressed" in manufacture. The disc then does not need any extra plating, etc. Looks like there's hope vet for audio enthusiasts who think they're in a rut.

And it's not only at JVC where all the action is. There's similar work going on in Poland where spies nform that at least one company out there is working on a capacative video disc player. The 32cm disc gives about 30 minutes of colour TV per side. It is interesting that recording techniques of this type are dividing. JVC, a company in Poland, and the mighty RCA are all working on similar systems, while another method using optical pick-up techniques is favoured by such people as Philips and French giant Thomson-CSF.

Wait for it

A new chip for hi-fi buffs will make constructors lives a little easier. Instead of all those expensive and bulky potentiometers (usually ganged, too) this chip will offer elegant and very compact control of base, treble, balance and volume. Direct-current control signal is the secret and can be effected by means of varying a voltage in the range 0 to 6V. The total gain variation is said to be of the order of -72dB to +12dB on both base and treble. It is hoped that the new chips will sell for £1-£2 although these are not as yet available. Until they do arrive, readers will just have to control themselves!

Hot Plasma

If you're the proud owner of a microwave oven you are not the only one putting those magic waves to good use. A short while back, an American University managed to heat some plasma to a mere 60,000,000 C. The interest in such fantastic temperatures is a thing called sustained nuclear fusion which is going to give us all a source of power when there's no oil or other fossil fuels. Snag is, that for sustained nuclear fusion you need things just a might hotter; like 100,000,000 C!

Enter another University complete with a "toroidally-shaped Tokamak chamber" (in Barnsley, they speak of nothing else). Into this luckless toroid, it is intended to beam microwave energy. Technical buffs hovering around the project are muttering something about needing to supply 4,000,000 Watts of power at 4GHz. One of the goals in nuclear fusion is to contain the plasma for a whole second at 100,000,000 C. Apparently this makes hydrogen atoms change, by fusion, into more complex atoms and release energy. Definitely not in the shops-yet.



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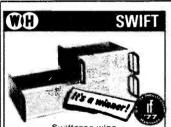
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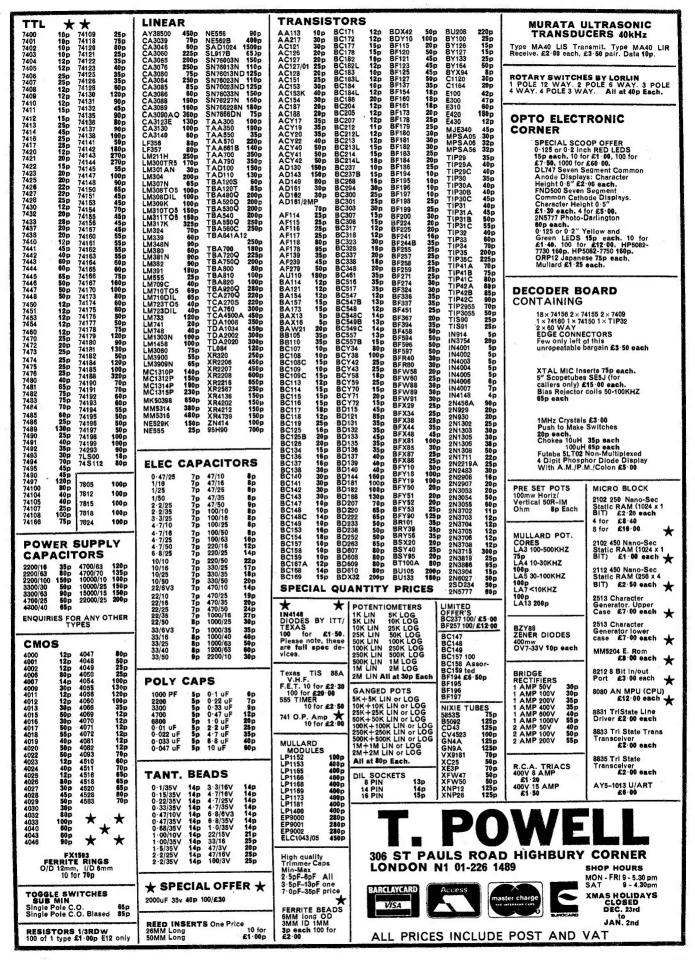
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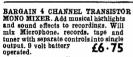
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Practical Wireless, January 1979



The conventional telephone system is essentially electrical in nature. The audio voice signal is converted into audio frequency electrical signals by means of a microphone before it passes to the local exchange; although the form of the signals may be changed in some way (for example, into a pulse code modulation signal or into a microwave beam for carrying the signal to a satellite and hence to another continent), in conventional systems the signal always travels by electrical signals or radio waves.

The installation of an experimental fibre glass link into the UK telephone network last year was therefore something of a revolution in telephone techniques. In essence, the telephone signal is used to modulate the infra-red radiation emitted by a gallium arsenide laser diode. This radiation is coupled into a length of glass fibre no thicker than a human hair, the fibre being constructed so that virtually all the radiation is reflected from the sides of the fibre along its length.

The length of such fibres may be of the order of 1km, but a few such fibres may be connected end-toend so that a single fibre-optic link may carry a signal a distance of the order of 10km. At the receiving end the infra-red radiation is fed to the junction of a sensitive infra-red detecting diode which converts the signal back into an electrical voltage or current. After amplification this signal can be fed through the telephone network to any destination.

Fibre Production

It is essential that optical fibres for telecommunications applications should be produced in long lengths. Although lengths of fibre can be joined, there are inevitably losses at each junction. Such losses can amount to as much as 1 to 2dB per junction, but the use of very refined jointing techniques can reduce the losses somewhat. Nevertheless, it is essential to keep the number of junctions to a minimum in order to obtain efficient communication.

In practice it has been found possible to manufacture glass fibres in 1km lengths for telecommunications work. It is essential to employ ultra-pure chemicals to manufacture the glass, since impurities can act as light scattering centres which introduce losses into the fibre. The losses in a typical fibre are of the order of 5dB per km at present, but fibres with considerably lower losses have been produced in the laboratory.

The structure of an optical fibre cable is shown in Fig. 1. Two steel wires are embedded in flattened

polythene strip to provide the required strength. The very small diameter optical fibre cables are placed in cavities in the polythene material. Graded optical fibres have been used in the Post Office experimental link; that is, the refractive index of the fibre decreases gradually with increasing radial distance from the centre of the fibre. This type of graded fibre results in the radiation being returned to the centre of the fibre not by total internal reflection from a sharply defined interface, but by a more gradual refraction in the outer layers of the glass fibre.

The use of such graded fibres reduces the spread in time which various rays take to pass through the fibre and enables a better performance to be obtained with a wide bandwidth. Fibres of graded refractive index can be made in long lengths by drawing glass simultaneously from two crucibles, the inner crucible containing the core glass and the outer crucible the cladding glass. The boundary diffuses as the glasses flow out of the crucible, thallium being employed as the diffusing ion.

The PO trial

The Post Office experimental trial of optical fibre communications has been used to link the PO Research Centre at Martlesham Heath with the main Ipswich telephone exchange, a distance of some 12km. Only one repeater was used, this being in the Kesgrave telephone exchange which is situated

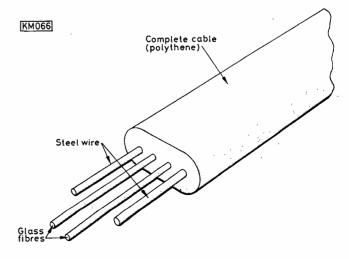


Fig. 1. The structure of the optical fibre cable used in the Post Office trials in East Anglia.

approximately half way between Martlesham Heath and Ipswich. Cable of the type shown in Fig. 1 has been employed, each glass fibre having a diameter of $62 \cdot 5 \mu m$ ($0 \cdot 0625 mm$). Two different types of cable have been tried, one type having a bandwidth of about 200MHz.km and the other a greater bandwidth of about 400MHz.km. In both types of cable the losses are some $4 \cdot 5 dB$ per km, whilst the weight of a lkm drum of the cable is some 50 kg.

A total of about 14km of the lower bandwidth cable has been employed to link Martlesham with Kesgrave and Kesgrave with Ipswich. The work with this cable involved a 8.448Mb/s data rate feasibility study.

The 6km length of the higher bandwidth cable was used to link Martlesham with the Kesgrave exchange at data rates of up to 140Mb/s. An additional length of 1km of this cable was used for experimental work whilst wound on a drum.

The cables were installed in existing underground telephone ducts, but great care had to be taken to prevent breakages of the glass fibre, since no spare lengths of cable were available. A dummy cable of similar structure was therefore placed in the ducts initially and was used to draw the valuable optical fibre cable into the ducts. Each 1km length of the cable was installed by placing the 1km drum near the centre of the section and drawing the cable first in one direction and then in the other so that the maximum length of each section to be inserted into the ducting did not exceed 500m.

Special tools were designed to enable low-loss joints to be made between successive sections of the cable. A tool with a tungsten carbide blade was used to "glue" the two ends together, the fixing of the ends The ends of the two cables to be joined must be very accurately aligned, since an alignment error can result in considerably greater signal losses.

The Post Office engineers employed a jointing machine to align the two ends of the cable. An epoxy adhesive of suitable refractive index was then used to "glue" the two ends together, the fixing of the ends being accomplished in a few minutes by raising the temperature of the epoxy adhesive.

Performance

The optical fibre experimental link connects four telephones at the Martlesham Research Centre to the Ipswich telephone exchange. Telephone calls can be directly dialled from any of these four telephones to most of the 22 million telephones in the British Isles. However, the optical fibre links used are capable of carrying up to 120 telephone calls simultaneously. The separate fibres each carry signals in one direction only.

An important feature of the system is that only one intermediate repeater (or signal amplifier station) is required, this being at Kesgrave. If the signals were sent for the same distance along conventional telephone lines, the signal would have to be boosted at six different points. The small diameter of the optical fibres will enable large numbers of them to be grouped together and installed in small diameter ducting so that many telephone calls can be carried simultaneously.

In the Post Office experimental link, data and clock inputs at 139.264Mb/s are taken to an encoder and the signal is processed so as to convert it into a form suitable for the current drive of a laser diode. There

is a coupling loss between the diode emitter and the cable of some 0.5dB, this leaving a maximum power of some $850\mu W$ to enter the fibre. The wavelength of 840nm is in the near infra-red region.

A digital "1" signal level corresponds to the $850\mu W$ level approximately. Owing to laser threshold delays, the digital "0" signal is not made the level at which no radiation is emitted from the laser diode, but rather a level of about $70\mu W$ so as to keep the diode in its lasing state at all times. The mean input power is thus some $460\mu W$, but the signal emerging at the far end of the line has a power level of only about $1\cdot4\mu W$. This output signal is coupled into an extremely sensitive avalanche photodiode which provides an input to an amplifier using a gallium arsenide MOSFET.

The maximum distance between repeater stations in the present system is some 9km at infra-red input power levels of 1mW. Calculations indicate that optical fibre systems can be designed which have a cost of only two thirds of that of conventional cable connections. It may well be that optical fibres will not be used to convey signals to individual houses and other premises, but rather to convey large numbers of multiplexed signals between telephone exchanges.

The length of optical fibre between Martlesham and Kesgrave, a distance of nearly 6km, shows a loss of some 25dB overall with a bandwidth of about 160MHz. Transmission of data is thus possible at rates of up to at least 300Mb/s.

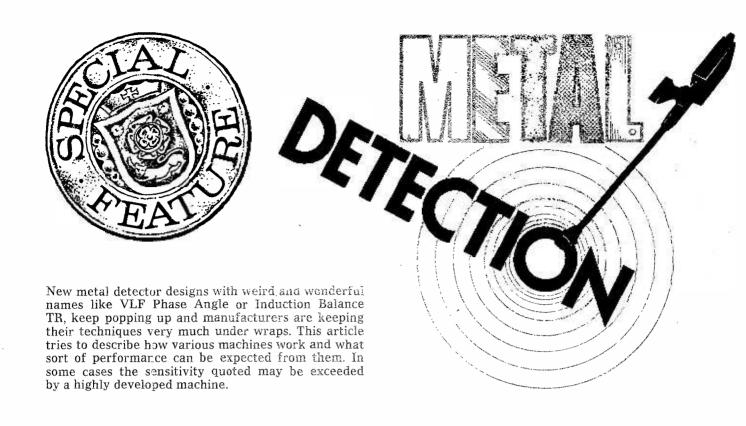
Conclusion

The Post Office field trials and other experimental links using glass fibres (such as the Standard Telephones and Cables link between Hitchin and Stevenage laid in Post Office telephone ducting) have shown that this form of communication is readily compatible with the existing telephone networks and it seems that it is likely to be very cost effective.

Although it is impossible to forecast future developments, one can imagine that optical fibres may be extremely attractive when one wishes to convey wide bandwidth signals rather than a narrow bandwidth telephone conversation.

Optical fibre communications could even make it an economic possibility to introduce the "videophone"—seeing by 'phone as well as speaking!





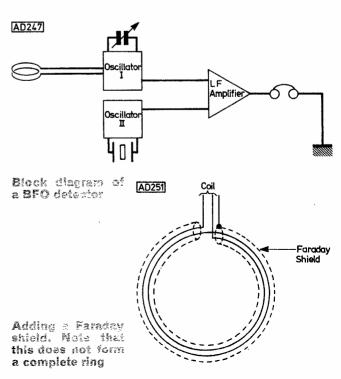
Beat Frequency Oscillator (BFO)

The BFO is the simplest type of metal detector and hence it is the least sensitive and the most difficult to use. The essence of the design is having two oscillators tuned to frequencies very close to each other. One oscillator is the reference and the other is tuned by the inductance of the search coil, which will alter as metal is brought near it. Either oscillator can be tuned by the operator, depending only upon the manufacturer's choice. When the two oscillators are at nearly the same frequency, mixing their outputs together produces a beat note which is equal to the difference between the two frequencies. This note is arranged to be in the audio spectrum, and thus it can be amplified and fed to the headset.

The output from the headset is thus a tone which increases or decreases in frequency when the coil is brought near to metal. If the metal is ferrous then the inductance of the coil will increase and the frequency of the oscillator will fall. This gives a falling note in the headset. The opposite applies to non-ferrous metals which gives a rising note in the headset. Needless to say, this can be reversed by tuning to the other side of the reference oscillator. Extra sensitivity can be obtained by tuning the reference oscillator to a frequency which is a multiple of the coil oscillator and using a phase-locked loop to control the system as in the PW "Ferret".

One disadvantage of these machines is that two oscillators are used. Each can drift causing constant retuning to be necessary, although some manufacturers have used a crystal-controlled reference leaving only one oscillator with a drift problem. These machines are also affected by ground capacitance causing the coil oscillator to vary in frequency, though this can be partially avoided by using a Fara-

day shield round part of the coil. The BFO machine cannot tune out silver paper, coke, ring pulls from beer cans etc. and will be affected by salt water, wet grass and terrain changes. Its typical sensitivity is 100 to 120mm on a 2p coin.





Induction Balance (IB) and Transmit/Receive (TR)

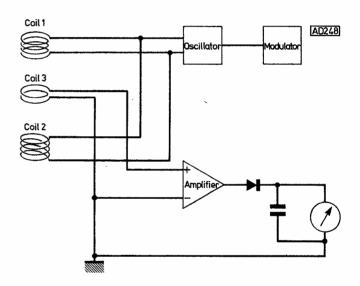
A great deal of confusion has arisen from these two names, one manufacturer calling his IB a TR and vice-versa and other manufacturers calling them IB/TR machines. They both work on the same principle of Induction Balance though even this is somewhat of a misnomer. The true IB machine has three coils, while the TR has two coils. The effect used for detection is the change in coupling which occurs as the coils are brought near to metal. Both machines use only one oscillator.

In the IB machine the oscillator is fed to two coils. The three coils are arranged on top of each other, very slightly displaced from centre, and the top and bottom coils are connected to the oscillator. The bottom coil is wound in the reverse direction to the top one, and thus the two coils generate antiphase fields. The third coil is placed between the other two where the fields balance and thus, at its terminals, has no signal. When metal is brought near to the coils an imbalance occurs and the centre coil then picks up a signal as the amplitudes of the other signals change unequally. The output from this coil can then be amplified and rectified to drive a meter, and the oscillator is usually amplitude modulated to produce an audio tone for the headset.

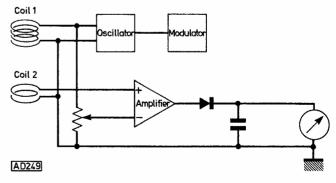
The TR machine has only one coil connected to the oscillator. The other is connected to the non-inverting input of an amplifier. The inverting input is connected via the control potentiometer to the oscillator and thus the amplifier can be set to give no output when the coil is not near metal. As the coil approaches metal a transformer effect occurs, coupling more signal into the receiving coil. The imbalance is amplified as before for the headset and meter.

In both machines the coupling increases for ferrous metals and decreases for non-ferrous, so by mixing the output from the amplifier with the signal from the oscillator the resulting signal will increase for ferrous metals and decrease for non-ferrous metals. It has also been found that small ferrous objects and aluminium objects, silver paper or bottle tops exhibit a skin effect where the coupling causes phase changes, and thus these items can be discriminated against. The final result is a comprehensive metal locator with lots of control knobs. This makes it look very impressive but unfortunately also makes it very difficult to set up perfectly. Generally, if the discriminating mode is used, the sensitivity of the machine is reduced.

The problem with both these types of machine is the ground effect, which causes changes in the coupling of the two coils and thus causes false readings if the coil is not held at a constant height above the ground. Salt water, wet grass and terrain changes can also upset the machine, although manufacturers are introducing Ground Exclusion Balance (GEB) machines which can eliminate any effect caused by the ground. GEB machines are expensive at around £250 and are complex to set up. A good IB/TR machine will detect a 2p coin at over 200mm.

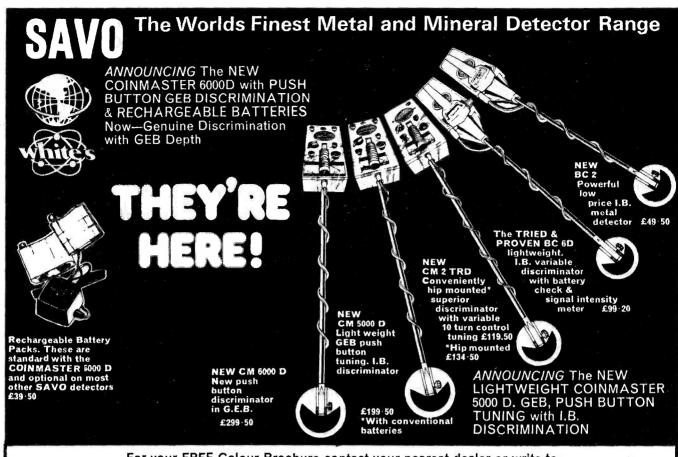


Block diagram of an IB detector. Coil 2 is wound in the reverse direction to Coil 1



Block diagram of a TR detector. The balancing signal is adjusted by means of the potentiometer





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Pulse Induction (PI)

The PI machine is described elsewhere in this issue, so it is unnecessary to spend too much time explaining how it works. The stability of the Pulse machine is vastly superior to any other and it does not suffer from ground effect. It cannot differentiate ferrous and non-ferrous materials and has a relatively high

power consumption. The other important factor is that the PI machine is an absolute metal detector, whereas the other machines are relative metal detectors. This means that on a relative metal detector it is possible to put metal near the coil and then tune it out without losing sensitivity. On the PI machine this is not possible, and sensitivity will be reduced if any metal is near the coil when the machine is set up. The net result of this is that PI machines cannot have metal shafts etc, but other machines can.





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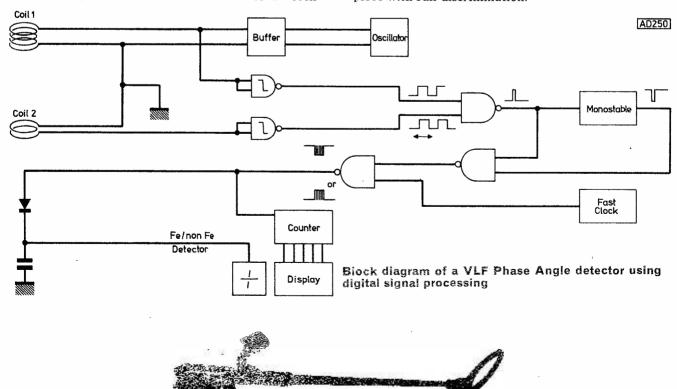
VLF Phase Angie

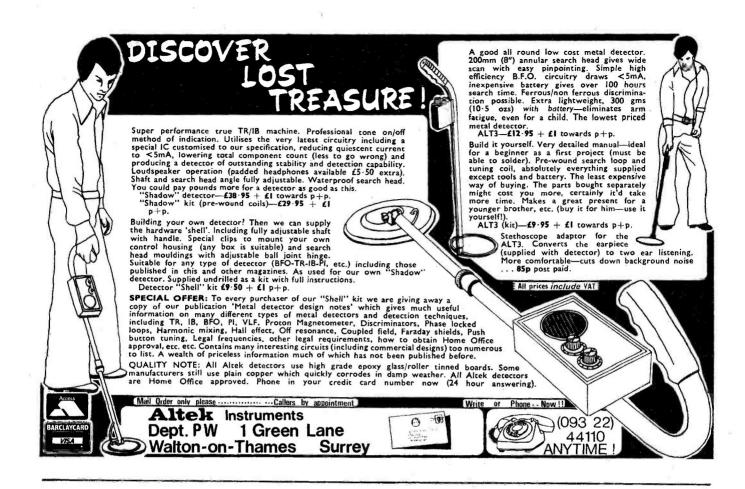
The VLF stands for Very Low Frequency and refers to the frequency of the oscillator used to drive the coil. The advantage of using low frequencies is to reduce the effect that the ground capacitance has upon the circuits, and thus the coils tend not to be affected by terrain changes. VLF is also used in IB machines which have the GEB suffix and it is this factor which contributes so much to their expense. When low frequencies are used the changes produced in the coils are much less than for high frequencies, no matter which technique is being used for the detector, and thus much greater signal processing and hence stability is necessary to produce a machine with the same capabilities.

The phase-angle method of metal location is the latest development in this field, and as yet is very underdeveloped by commercial manufacturers. It offers the greatest scope in location techniques as it can easily be made with digital electronics. and thus a great deal of digital signal processing can be carried out at relatively low cost. The phase-angle machine has two coils, one of which is connected to an oscil-

lator via a buffer stage. This ensures that the coil cannot alter the frequency of the oscillator. The second coil is connected to an amplifier which produces a square wave at the same frequency as the oscillator. The oscillator is also fed into an amplifier to produce another square wave. When the coil is brought near to metal, the phase difference between the two square waves alters, increasing for ferrous and decreasing for non-ferrous. The implications of this phenomenon should now be obvious. The simplest form the circuitry could take would be to invert one signal and add the results to produce a spike. Feed this into a smoothing circuit and there you have a voltage to drive a meter. To improve the gain, a monostable could be used to generate an inverted spike of the same period to cancel out the first spike. Followed again by a smoothing circuit and now a d.c. amplifier and the gain is vastly improved.

To continue in the digital sphere, a fast clock could be gated by the spike to produce a count for a digital readout which could read + for non-ferrous and - for ferrous, with the magnitude of the number indicating the depth and size of the find. Sensitivity on these machines can easily be 250-300mm on a 2p piece with full discrimination.





Licensing Requirements

Having, hopefully, provided food for thought on designing and building your own metal locator, or even for the wealthy, going out and buying one, there are other matters to be considered. The first point is that to use one of these machines a licence is required. It is referred to as a Pipe Finders Licence and costs £1·40 for 5 years, application forms are available from the Home Office, Radio Regulatory Department, Waterloo Bridge House, Waterloo Road, London SE1 8UA. A licence covers only one type of detector, and if you buy another design you need a new licence.

The Home Office also has regulations concerning the signal in the coil and if you are designing your own equipment you should stay within their specifications which state:

1. Air-cored or ferrite-cored coils may be used but the energisation shall not exceed 0·1 amperes r.m.s. per turn.

Modulation, either a.m. or f.m. may be used if required.

3. The depth of a.m. shall not exceed 100% so do not use square wayes.

4. The authorised bands for use are 0-16kHz and 16-150kHz.

In a nutshell, they say do not use a 2N3055 to switch the coil across a car battery at 1MHz!

General hints on designing your own metal detector are: try to keep the operating frequency as low as possible. This reduces the effects of ground and water. Frequencies below 25kHz should produce few false readings from the terrain.

An audio output gives far more dynamic range to the man-machine interface than a meter, and the changing pitch of a voltage-controlled oscillator (v.c.o.) is far better than a rising volume output. Do not worry about driving a loudspeaker, a headset is



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The Pioneer TRIB utilises the same principle as the Viking but has the following additions: Extendable Shaft, 7" Swivel Search Head and is of ultra lightweight construction. The Invictus D TRIB Discriminator is one of the lowest priced discriminators on the market today, sacrificing none of its performance against detectors costing much more.

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far easier to use and makes it easier to detect minute changes in the signal in high ambient noise levels. When developing your circuit, pay attention to supply decoupling as large currents in coils or headphones can cause most weird effects on the sensitivity of your circuit. The size of the coil is important both for inductance and sensitivity. The larger the coil the more sensitive the machine, but the more difficult it becomes to find the centre of any metallic object discovered.



Using your Detector

Building or buying a metal detector is one thing but using it is something else. Unless you are on your own land you need permission to use one, and in many parts of the country the use of metal detectors on public land (parks, footpaths etc.) is forbidden by a bye-law, so check first.

The other main rule is that which concerns any finds that you may make. Gold and silver are classed as Treasure Trove and are the property of the Crown, and must therefore be handed in to a Coroner, who will hold an Inquest to decide true ownership rights. Items of historical value usually end up in a museum and in order to promote handing in of Treasure Trove, the finder is generally given a reward to the value of the items. Further information on Treasure Trove may be obtained from the Director, British

Museum, London WC1. Any other items found become the property of the owner of the land, not the finder, so it is best to come to an agreement with the landowner before you start searching. Archaeologists do not like metal detectors or people who use them so do not go on known historical sites with yours. You could be prosecuted!

After you have dug up everything in sight and found very little, you may care to do a little research at your local library. Establishing the likely positions of ancient Roman camp sites may not be easy, but professional metal detector users, and there are many, spend nine months researching and three months out in the field with their machines. One tip is to ask your local museum curator where not to dig, and if he gets his atlas out then try another museum. Curators will generally be found to be very helpful about known sites which are taboo and possible sites which are likely.



Waving your machine at the cows





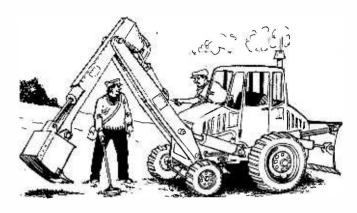
Having built or bought your machine, and researched your possible site, and got permission from the owners of the land, how do you find large hoards of gold? If I knew I would not be writing this article. It is no use walking aimlessly around waving your machine at the cows to frighten them away, and to expect to trip over a large treasure chest. Take some pegs and some string with you and mark out a square about 3m side. This is the first area to search and it must be searched fully.

If you have a BFO or IB/TR machine without GEB, you will find that the height of the coil above the ground is critical. If you set up your machine with the coil resting on the ground then as you raise the coil to start searching it will give a reading. If the machine is set up in the air, then when it is placed near the ground it will become less sensitive. The correct way to set them up is by adjusting the controls with the machine on the ground and then backing off the sensitivity just a fraction so that when searching the tone is just audible. As you move around, signals will be heard as the coil moves up and down, and only practice will allow you to know which is a find and which is not. When you sweep, the action must be slow and smooth and that comment applies to PI and Phase Angle machines as well. Integrators take time to work.

Having searched the area in your 3m square completely, lift two pegs and move them to form another



Trip over a large treasure chest



Use a small trowel, not a JCB!

square adjacent to the first. In this way it is possible to be sure that the whole of the suspect area has been searched. Remember that every centimetre the coil is above the ground is a centimetre it cannot detect below ground, so keep the coil low down. When the machine has detected something, concentrate on finding exactly where the object is. Sweep the coil NS and EW to find the exact centre, raising the coil head and retuning if necessary to reduce the sensitivity of the machine. Dig your find up carefully, using a small trowel, not a JCB! Remove a triangle of turf and then the minimum of earth, checking frequently to see whether you have dug it up or if it is still there. Do not throw scrap metal finds away; you may find them three or four times again if you just chuck them about, take them with you. Replace the earth that you have removed and stamp the turf down evenly so that it is impossible to see where you dug. Do not leave a trail of potholes and divots behind you. Good hunting!



Introduction to

A cursory glance through any catalogue of digital integrated circuits will show that there are some hundreds of different logic devices available. Many of these devices will simply be various arrangements of the basic gate or flip-flop elements, and there may be one to four of these logic elements within a single circuit package. Typically there might be some twenty to fifty transistors etched and diffused into a single silicon chip to form such a device. Such simple logic elements are usually referred to as Small Scale Integrated or SSI circuits and they will generally be packaged into a 14- or 16-pin DIL assembly. The 7400 and 4011 gates and 7474 or 4013 flip-flops are in this logic category.

As techniques for fabricating integrated circuits improved it became possible to build more complex circuits, having perhaps a few hundred transistors, on to a single silicon chip. These more complex logic devices are called Medium Scale Integrated, or MSI, circuits and most of the newer devices in the TTL and CMOS ranges tend to be of this type. Now it becomes possible to have the four flip-flops and the associated gates for a complete decade of a BCD counter as a single logic unit. Examples of this are the 74160 and 74190 in TTL or the 4518 in CMOS. In fact the latter unit contains two complete decades of a BCD counter. Similarly complex gate arrays such as those for decoding BCD inputs to produce ten-line or seven-line outputs for driving a display can be manufactured on a single chip.

What are the advantages of MSI circuits in comparison with the more basic SSI elements? First the number of integrated circuits needed for a given logic system will be reduced, since some five or six SSI devices are replaced by each MSI circuit that is used. As a result both the size and complexity of the circuit board can be reduced and a more compact logic

system can be produced.

When an integrated circuit is manufactured the testing and packaging of the device form a significant part of the total manufacturing cost whether it be an SSI or MSI device. As a result of this the MSI circuit will often cost less than the equivalent set of SSI devices. Another advantage is that there will be less connections to be made. For the amateur this means a saving in time and less chance that an error in wiring will be made. Reliability tends to improve since there are fewer soldered joints in the system. In the device itself the interconnections are produced by plating and etching which are inherently more reliable than solder for making connections.

Although MSI has its advantages it does not completely eliminate the need for the simpler SSI devices. In general MSI circuits are used for the more common logic sub-systems, such as counters, registers and complex gate arrays, but SSI devices are still required for the interconnecting logic and for any special logic functions that may be required.

Logic Families

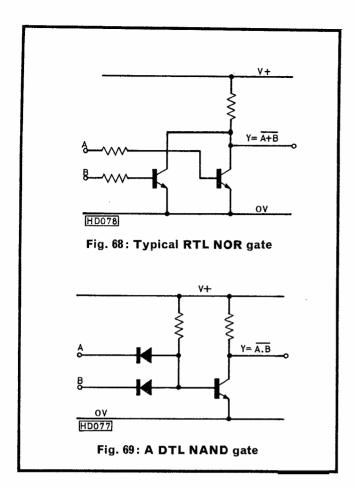
Apart from the TTL and CMOS type logic elements that we have been discussing in this series there are a number of other families of logic elements which may be encountered. Although they follow the same basic rules of logic, an AND gate still has the same logical action, the signal levels and the input/output characteristics will vary from one logic family to another. It may be interesting to look at the characteristics of some of these other logic types and see how they compare with TTL and CMOS.

One of the earliest types of logic device to be made in integrated circuits used Resistor Transistor Logic (RTL) where the gate elements consisted entirely of resistors and transistors. Typically a 2-input NOR gate would have the arrangement shown in Fig. 68. If either input is driven to the 1 level, its associated transistor will turn on and pull the output line down to the 0 level. RTL devices were generally run from a supply line of about +3 to +3.5 volts and tended to be relatively slow in operation. This type of logic has now virtually disappeared from the scene.

Prior to the arrival of TTL one fairly popular type of logic was Diode Transistor Logic (DTL) which is still available today although it is steadily being replaced by CMOS devices. Here the gates are basically diode gates as described in the early part of this series and these are followed by transistor buffer stages to provide the output drive capability. A typical DTL NAND gate would have the circuit shown in Fig. 69. There were several variations of DTL made by different manufacturers but the most common were the 900 series devices. Typical of these was the 930 which contained two 4-input NAND gates. Logic signal levels and supply voltage are basically the same as for TTL but the switching speed is about three times slower than TTL. In fact DTL and TTL devices can be used together, although a DTL gate can only drive about one TTL input.

Low Power Logic

A number of variations of the 7400 series TTL devices have appeared over the years since TTL was first introduced. One of the earliest of these was Low Power Transistor Transistor Logic (LPTTL) in which the internal resistors of the logic element are increased to reduce the supply current. These low power versions of TTL also used 7400 series type numbers but an L was included to denote a low power



type. Thus the 7400 becomes a 74L00 in its low power version but otherwise the function and pin layout are the same. It should be noted that one or two 74L devices do not have the same pin layout as their 74 series equivalents so it is as well to check this before using a 74L to replace a 74 device.

Typically a 74L series device will consume a quarter of the current that a normal 74 device does but this has to be paid for by switching speed reduction to about a third that of a 74 series device. Normally an LPTTL circuit will be capable of driving some two standard TTL loads. This type of logic is useful for battery-operated systems, although it is now tending to be replaced by CMOS which is less critical on supply voltages.

One interesting variation with 74 series numbers is the 74C00 type which is not TTL at all but uses CMOS logic. In fact these are CMOS devices designed to have the same pin layout and function as equivalent numbered 7400 series logic.

High Speed Logic

Normal TTL devices are limited in speed to about 30MHz because the transistors in the gate or flip-flop are run into saturation when they switch. Excessive base drive will build up a charge on the base which has to be removed before the transistor can switch off and this causes the slower switching action. One form of TTL element which overcomes this problem is Schottky TTL whose devices have 74S series numbers. In a Schottky TTL circuit a Schottky diode is connected between the base and collector of all switching transistor stages as shown in Fig. 70. This type of diode has a low forward voltage drop, which

allows excess current in the base circuit to flow directly to the collector via the diode thus preventing the transistor from going fully into saturation. Now there is little charge to be removed from the base and the stage switches off more rapidly.

A typical Schottky TTL device will be about three times as fast as its normal 74 series counterpart but it will require a little more supply power. With Schottky TTL more care is needed with decoupling and circuit layout if reliable operation is to be achieved.

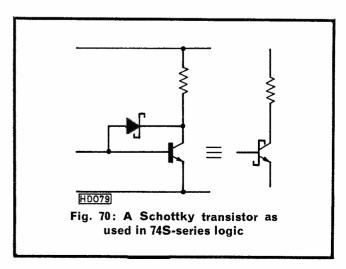
For even higher switching speeds, up to perhaps 600MHz, Emitter Coupled Logic (ECL) devices are used. In this type of logic the transistors do not saturate at all thus giving a high switching speed. Standard ECL devices such as the Motorola 10000 series are used in computers but are not normally available to the amateur user. Some special ECL devices such as divide-by-ten counters which operate at u.h.f. (made by Fairchild and Plessey) can be obtained however, and may be useful for v.h.f. frequency counters or synthesisers. These devices generally run from a -5V supply and are not directly compatible with TTL circuits.

There is also a low power version of Schottky TTL which has type numbers starting with 74LS. These devices have about the same switching speed as standard 7400 TTL but draw about one fifth the amount of power. Output drive capability is about the same as for 74L devices.

Logic System Design

We have seen how the various logic elements work and how they can be used but one question remains to be answered. How is a logic system designed? The first step in the design process is to decide exactly what the logic system will be required to do. This may seem obvious and yet it is quite surprising how many engineers try to design logic systems without first getting absolutely clear in their minds what the logic has to do. This basic rule applies not only to the design of new systems from scratch but also to modification of existing logic systems to make them do something different. In the latter case it is, of course, helpful if the action of the existing system is clearly understood before any attempt is made to alter it.

For any logic system there will be a set of input signals and some output signals and these will be related in some way to one another. Also there may



be a sequence of actions that the logic must carry out or events to which it must respond. In a simple system, once these things have been defined, the actual logic arrangement can often be worked out by simple commonsense reasoning.

As an example, let us suppose that we have to design a warning system that will alert a car driver to the fact that his car lights are still turned on after he has turned off the ignition. Here we have two input signals, one from the ignition switch (I) and the other from the light switch (L). The output of the logic, which we shall call A, is used to operate an alarm buzzer when I=0 (ignition off) AND L=1 (lights on). We can write this down as;

$$A = \bar{I}.L$$

Now this looks as if it could be met by using an AND gate. However to get the proper action we shall need to invert the I input signals so that both inputs of the AND gate go to l when the alarm condition exists at the input. The alarm must not operate if l=l or l=0. The circuit shown in Fig. 71 will perform the required function.

There are various other gate arrangements which will also meet the requirements of this alarm system. How about using a NOR gate instead of the AND arrangement? Think of what you must do to the inputs to make such a scheme work.

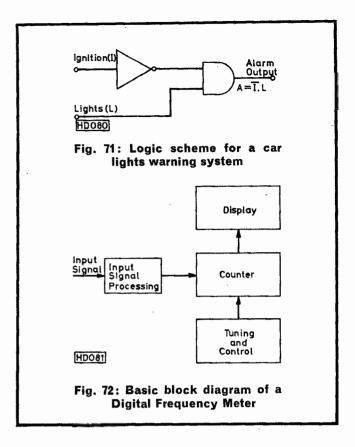
Digital Frequency Meter

When more complex systems are designed, it is usual to try to break them down into a series of subsystems each of which can be dealt with separately. In a very complex arrangement even the sub-systems may be broken down further into smaller sections of logic. Having decided what the main system must do the action of each of the sub-systems can be defined and they can be designed as independent systems. At all times however the interaction between the subsystem must be borne in mind since if the action of one sub-system is altered it may affect the design of some of the others.

As an example of a more complex logic system let us consider the design of a digital frequency meter. How can we measure frequency digitally? In the old days, before the term hertz crept into our language, frequency was measured in cycles per second, or maybe kc/s or Mc/s for the higher frequencies. Suppose we count up the number of cycles of the input signal that occur in a period of exactly one second. The resultant total will be the frequency in cycles/second or in our modern terminology it will give the frequency in hertz.

At the heart of our frequency meter therefore we shall have a counter to total up the number of cycles of the input signal. Having obtained this total we shall need some kind of display system to provide a numerical readout of the frequency we have just measured. Next we need an accurate clock to determine the one-second period over which input cycles will be counted, and some control logic to regulate the sequence of events involved in making each measurement of frequency. Here the counter must first of all be reset to zero, then it must count input cycles for some fixed time period and finally the result must be presented on the display.

The input signal itself may need a certain amount of processing to make it compatible with the logic system. Usually this simply consists of amplification



and perhaps shaping of the waveform to give squareedged logic signals. So we end up with a basic system diagram which is roughly as shown in Fig. 72. The counter, display, timing and input sub-systems can now be dealt with separately to arrive at the detailed logic design, and we can look at the options available in the design of these sub-systems.

The Main Counter

Starting at the heart of the unit, let us consider the counter which will total up the number of cycles of input signal. The simplest form of counter would be a pure binary type. For convenience in use the display will need to be in decimal form and this is not normally compatible with a pure binary counter output. The obvious choice here is to use a BCD counter chain to total up the cycles of input signal. Matching the BCD outputs to a decimal display system then becomes quite straightforward.

How many decades of counter de we need? The answer to this question will depend upon the range of frequency to be covered and the precision to which frequency is measured. If we want to cover the range 0 to 30MHz in 1kHz steps then a counter reading from 0 to 30,000 is needed so there will have to be five decades in all.

What kind of logic is needed? In order to achieve a counting rate of 30MHz either TTL or Schottky TTL might be used. Even at 15V supply, CMOS devices would not be capable of this speed. However, CMOS could be used if we build some form of prescaler counter into the input signal processing circuits. Suppose we divided the input frequency by ten before making our measurement. Now the counter would have to work at only 3MHz and CMOS could cope with this easily. In order to get the correct answers, however, we would need to increase the time period over

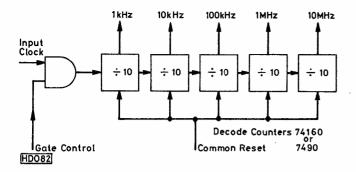


Fig. 73: A typical scheme for the main counter

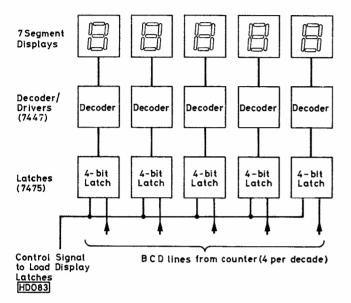


Fig. 74: Logic for the display subsystem

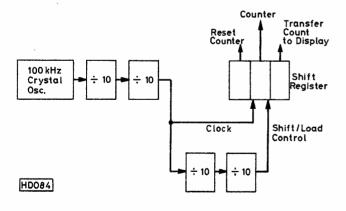


Fig. 75: Timing and control logic

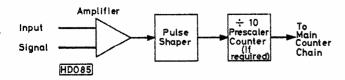


Fig. 76: Input signal processing

which the counter operates by a factor of ten in order to obtain the same total number of cycles. This prescaler approach is particularly useful if the signal to be measured is in the v.h.f. or u.h.f. range.

A simple AND gate ahead of the main counter chain is used to control the passage of the input signal pulses. When the counter is totalling up the cycles this gate will be open allowing the input pulses through to the clock input of the first stage of the counter chain. At the end of the counting period the gate closes and the count action stops.

The only other requirement for the counter chain is a common reset line to all stages of the counter to allow it to be reset to zero at the start of each measurement cycle. The resultant logic scheme for this main counter chain might be as shown in Fig. 73.

Display System

Either l.e.d., l.c.d. or Nixie tube displays could be used to provide the frequency readout. The choice will be governed largely by personal preference on the part of the designer or perhaps by the availability of a particular type of display unit. For battery-operated equipment the l.c.d. type display, with its low power consumption, has many advantages.

Apart from the displays, a set of decoder/driver logic will be needed to convert the BCD outputs from the counter chain into signals suitable for driving the display. Here the type of device required is governed by the type of display being used.

Display multiplexing could be used but with only five digits it may be found that there is no useful saving in the amount of logic required so the simple scheme shown in Fig. 74 might well be used. A latch has been included for each decade. If this were not used the display would flicker as the counter went through its counting operation.

The choice of logic type for the display system will be governed mainly by the capability of the devices to drive the displays. Speed of operation is relatively low, so all of the logic types will be suitable in this respect.

Timing and Control

To measure off the time period over which counting is carried out, an accurate clock signal is needed. This is usually produced by a crystal oscillator running at 1MHz or perhaps 100kHz. The crystal frequency is divided down in a decade divider to produce clock pulses at say 1ms, 10ms or maybe 1 second intervals and these pulses will control the measurement sequence.

The control logic needed to implement the measurement sequence can readily be produced by using a shift register with a single *1*-state travelling along it. The outputs of the successive stages of the register can then be used to control the various steps in the measurement sequence.

At the start of the cycle a 1 is loaded into the first stage of the shift register and the output from this stage is used to reset the counter chain. On the next timing clock pulse the 1 state moves to the second stage of the register and the output from this stage opens the counter gate to allow counting to occur. One time period later, counting is stopped as the gate closes and the 1 that has moved into the third stage of the shift register is used to transfer the number

Continued on page 71



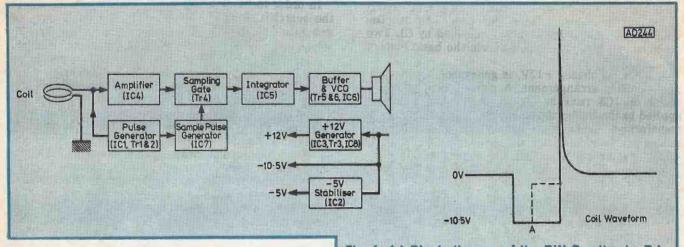
METAL DETECTOR P.J.Wales

Pulse Induction metal detectors have been available on the commercial market for some years and essentially their circuits remain unchanged from the original design. They have developed their own unique reputation because they have inherent features that other metal detectors do not have. They have always been very powerful machines capable of detecting a 2p coin up to 230mm underground. Because of the very low frequencies used they are insensitive to ground effects, coke, salt water, wet grass or small pieces of aluminium foil. Their sensitivity is a function of their current consumption and thus their battery life is generally poor, but their main disadvantage is their inability to discriminate between ferrous and non-ferrous materials.

The author has spent two years developing a circuit which combines good sensitivity, low current consumption and ease of use; a circuit which the enthusiast can build himself without the need for elaborate and complex test gear. The circuits have been particularly tuned to gold and silver, and they can be set up perfectly with only a meter. The unit illustrated is assembled into a case available as a kit

from Ambit International Ltd.





Theory

The basic principles of the PI metal detector are best understood by reference to the block diagram, Fig. 1(a).

As with most metal detectors, the heart of the machine is the coil. When power is applied to the coil it generates a magnetic field proportional to the number of turns of wire and the current passing through them. When the power is removed the voltage across the coil first drops to zero and then, as the magnetic field decays, builds up in the reverse direction as a back e.m.f. is induced into the coil. If the coil is correctly damped the back e.m.f. dies away as shown in Fig. 1(b). Should there be a piece of metal near the coil, it is affected by the magnetic field either to produce eddy currents or to be magnetised, depending upon whether the metal is non-ferrous or ferrous. In either case the result is the same in that the back e.m.f. takes longer to die away. This effect is most apparent in the time it takes to reach 0V.

Thus the detector circuits are arranged to process the area of signal where the back e.m.f. nears 0V. This is done by amplifying the coil waveform and then taking a sample of the voltage just as it gets to 0V. This sample is fed to an integrator which produces an output voltage proportional to the back e.m.f. decay time. This voltage is used to control the frequency of a voltage controlled oscillator (v.c.o.) which drives the speaker.

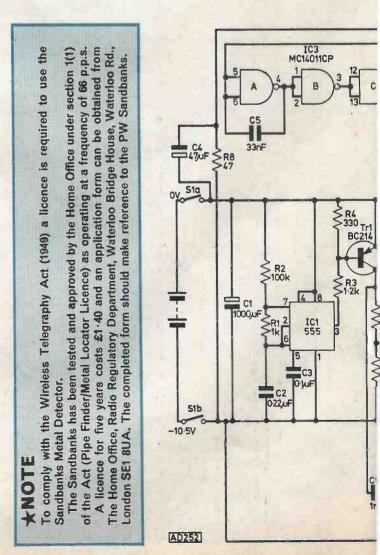
Circuit Description

The short negative-going pluses which are applied to the coil are generated by means of a 555 timer (IC1) operating in the astable mode. The repetition frequency is about 66 pulses per second and the duty cycle (ratio of pulse length to pulse interval) about one per cent. The output at pin 3 is amplified and inverted by Tr1 and applied to the power transistor Tr2. This transistor is run in a linear mode to increase switching speed and also reduce current consumption. The waveform on its collector follows the dotted line in Fig. 1(b) which shows that up to point A, Tr2 is saturated whilst the current builds up in the coil. After point A, the transistor goes into its linear mode, conducting approximately 0.5 amp.

Before describing the signal processing it is convenient to deal with the power supply arrangements. A 10.5V battery (seven cells) is used in the design illustrated, though any supply between 9V and 12V is satisfactory. The quiescent current consumption is about 50mA, rising to about 100mA on full detect.

Fig. 1: (a) Block diagram of the PW Sandbanks P.I. metal detector. Except for the power supply arrangements and component references, this is typical of any P.I. machine. (b) Coil waveform

Fig. 2: Complete circuit diagram of the PW Sandbanks



The positive line from the battery has been designated 0V. The other line is therefore -10.5V. Decoupling of the battery supply is provided by C1. Two other supply lines are derived from the basic battery supply.

The first of these, +12V, is generated by an oscillator-rectifier arrangement. A c.m.o.s. ring oscillator based on IC3 runs at about 30kHz. The output is applied to the buffer transistor Tr3 which has an autotransformer T1 as its collector load. The 50V peak pulses produced are rectified by D1 and smoothed by C6. Regulator IC8 maintains the output at 12V.

It is essential that IC3 is an unbuffered "A" type, since the buffered "B" type will not work in this circuit. No attempt is made to tune T1, as tuning tends to upset the ring oscillator, making it unstable. The circuit is decoupled by R8 and C4, to prevent 30kHz ripple being fed into the amplifiers.

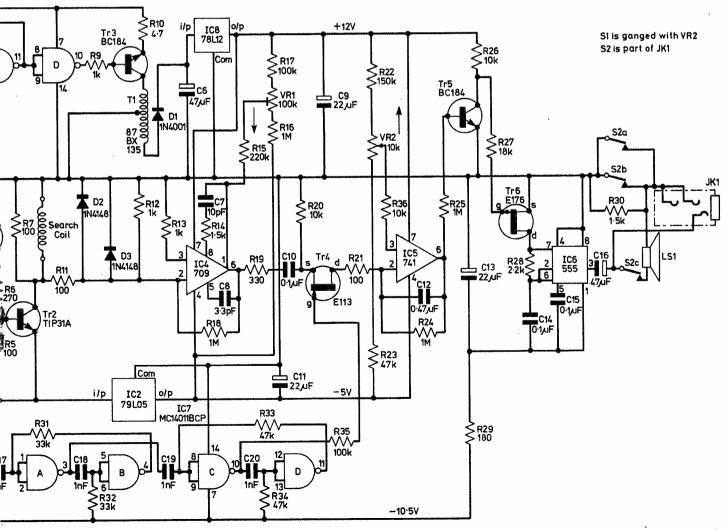
The second derived supply line is a stabilised -5V for the amplifiers. This is provided by regulator IC2.

The signal across the search coil is fed via a clipping network R11, D2 and D3, which limits the voltage swing at the input to amplifier IC4 to ± 0.7 V, to prevent overloading. The gain of IC4 is set to 10,000 by means of R18, and frequency compensation is designed to provide the fastest response with maximum reliability. The output offset is set by means of VR1.

In order to be able to sample the waveform across the search coil as it reaches 0V, it is necessary to generate a delayed gating pulse. This is done by IC7, which is arranged to form two monostables. The first, IC7a and b, produces a pulse $50\mu s$ wide, starting on the trailing edge of the pulse at Tr1 collector. The second monostable, IC7c and d, is triggered by the first, and generates a pulse $75\mu s$ wide, which is the sample pulse. This is applied to the gate of Tr4, turning the latter on for $75\mu s$, $50\mu s$ after the transmit pulse ends. See Fig. 3.

When the sample pulse is on the gate of Tr4, the source-drain impedance is very low, and the voltage at the output of IC4 is transferred to the input of IC5. For the rest of the period the impedance of Tr4 is very high, and IC4 output is isolated from IC5. The small negative-going pulses are amplified and integrated by IC5/C12 to form a low-rise sawtooth waveform whose d.c. level is proportional to the width of the back-e.m.f. pulse. The output offset of IC5 is adjusted by means of VR2.

The output of IC5 is connected to Tr5 which is a low-gain inverter. When IC5 output is low, the collector of Tr5 is at +12V and Tr6 source-drain impedance is too high for capacitor C14 to charge. When the output of IC5 is high, Tr5 is switched on and the source-drain impedance of Tr6 is low. This allows C14 to charge and IC6 then functions as an



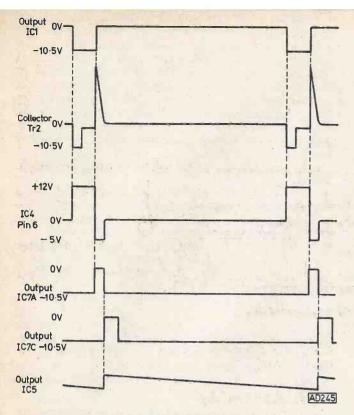


Fig. 3: Timing diagram

audio-frequency oscillator whose frequency is proportional to the amount by which the width of the back-e.m.f. pulse exceeds 50 µs.

The oscillator output is applied to the internal loudspeaker, or to headphones when these are plugged into JK1.

Construction

The battery holder should first be assembled sufficiently to provide power for testing. It will be finished when the circuits are built and tested. The top adaptor should be gently tapped into the aluminium tube using a piece of wood as a buffer to protect the pvc from damage. The spade terminal which forms the negative battery connection should then be rivetted to the tube with the special rivet provided. Remove the epoxy coating from the first 15mm of the tube with a Surform tool or file and glue the battery cap holder to the tube with an isocyanoacrylate adhesive such as Cyanolit or Super Glue 3. Stretch the spring so that it is about 50 per cent longer than its original length. It should now slide easily down the tube. Bend the contact plate extensions down at right angles and clip it over one end of the spring. Fold the extensions over the spring to clamp it firmly. Bend out the last 10mm of the spring to cause it to scrape the side of the tube and push it down the tube with the batteries.

The coil is wound from 20 turns of 26 s.w.g. enamelled copper wire in a 200mm diameter circle. The coils should be secured with twists of copper wire. ensuring that they do not form a short circuit around the coil. Do not use adhesive tape for this as the epoxy resin tends to lift it away and then the coils move whilst they are setting. Feed a 2m length of 3-core

* components

THE REAL PROPERTY.	
1	R10
1	R8
4	R5, 7, 11, 21
1	R29
1	R6
2	R4, 19
4	R1, 9, 12, 13
1	R3
2	R14, 30
	R28
	R20, 26, 36
	R20, 20, 30
	R31, 32
	R23, 33, 34
Service Control	R2, 17, 35
	R22
	R15
4	R16, 18, 24, 25
1	VR1
	E. T. S. C.
1	VR2/S1
William .	CONTRACTOR OF THE
1	C7
4	C17, 18, 19, 20
1	C8
TRIF	
1	C5
	C3, 10, 14, 15
	C2 C12
	CIZ
2	CO 11 13
3	C9, 11, 13
ating	
	CA 6 16
	C4, 6, 16
	C1
1	D1
2	D2, 3
	DZ, 0
	THE RESERVE OF THE PARTY OF THE
2	Tr3, 5
1	Tr1
1	Tr4 (or BF256B)
1	Tr6
1	Tr2 (or BD535)
-	
VV-15	
1	IC3 (see text)
1	IC7
1	IC8
1	1C2
2	IC1, 6
1	IC4
1	1C5
	1 4 1 1 1 2 4 1 2 1 3 1 2 3 3 1 1 4 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1

T1 Toko 87BX135. JK1 Switched stereo jack, PL1/SK1 3-pole connector, Bulgin P632. LS1 8Ω 2 or 2½in. Materials for case, battery holder, shaft, search coil etc. (see text).

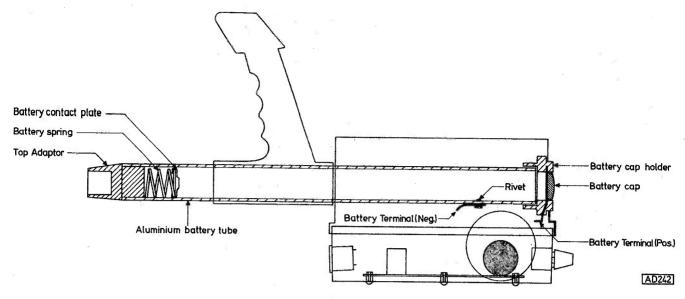


Fig. 4: Details of the case assembly

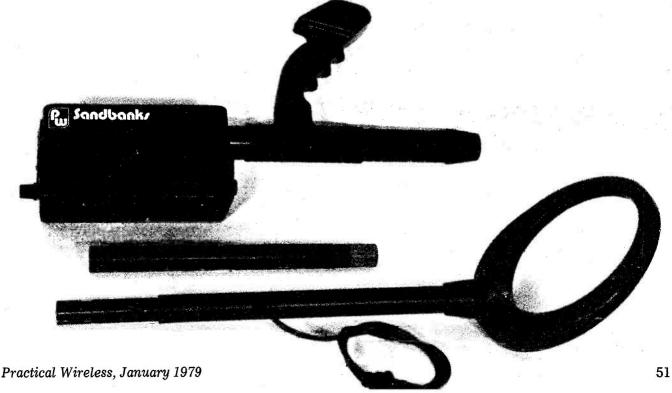
miniature mains cable into the hole in the side of the bottom tube and out of the machined end. Strip it and solder the blue and the brown to the ends of the coil. Glue the bottom adaptor into the coil with Cyanolit and leave it to set hard.

Place the coil in the moulding and shape it to lie as flat along the bottom as possible, ensuring the minimum resin used and the lightest coil. Seal the wire into the adaptor with Seccomastic to prevent the resin from leaking past the wire. Chock up the moulding until the top surface is horizontal and ensure that it is firmly supported. The epoxy resin in the kit has resin and hardener in one plastic bag, separated by a plastic clip. Remove the clip and mix thoroughly for at least five minutes. Snip off the corner of the bag and carefully pour it over the coil using only enough to cover it. If any resin is spilled, do not wipe it off the moulding as it comes off easily when it is set. Leave to harden for 24 hours minimum. An

alternative printed circuit coil is available from Ambit International.

PCB Assembly

Very few of the components are critical but to avoid problems it is advisable to use only new, best quality parts. The circuit has been designed so that a working unit can be set up perfectly using only an Avometer. Due to the complexity of the circuit, it is recommended that each stage is built and tested before progressing to the next. Accordingly a loading sequence is given for each part of the circuit followed by a testing and fault finding guide. Each part of the circuit must be working correctly before progressing to the next stage. It is inadvisable to try to build this circuit on Veroboard as the layout is fairly critical, and at least a double-beam 'scope will be necessary for fault finding if problems occur.



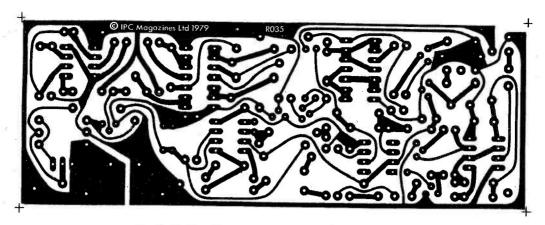


Fig. 5: Foil pattern of the p.c.b., shown full-size

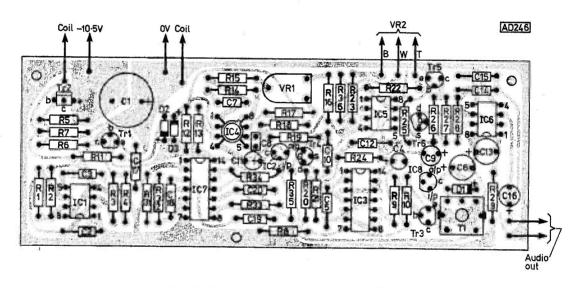
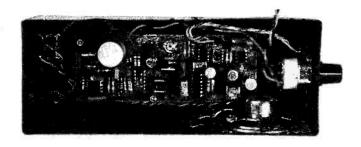


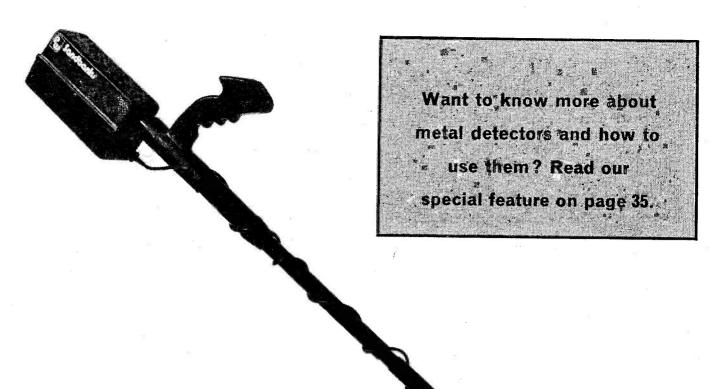
Fig. 6: Component layout on the p.c.b.



Pulse Generator: Load IC1, C1, C2, C3, R1 and R2 Check, crop and solder. Wire the p.c.b. to the battery spade sockets via S1 (part of VR2). Connect the spade sockets to the battery holder terminals. Switch on and measure the output at pin 3 IC1. It should be one per cent of 10.5V, i.e. 105mV. Switch off.

Power Stage: Load R3, R4, R5, R6, R7, Tr1 and Tr2. Check, crop and solder. Switch on and measure the voltage between Tr1 collector and the negative rail.

It should be the same as IC1 output, i.e. 105 mV. Switch off and solder the coil temporarily into the p.c.b. Make up an improvised peak-reading probe for the meter by wiring a 1N4001 diode between the positive test clip and its lead (positive or white end of the diode towards the lead) and a $0.1 \mu F$ capacitor across the leads. Switch on again and measure the voltage between the collector of Tr2 and 0V (meter negative to 0V). This can be anything between 12V and 50V, this being the peak back e.m.f. voltage. Switch off.



Minus 12V Supply: Load R8, C4, C5, IC3, R9, R10, Tr3, D1, C6, IC8, C9 and T1. The usual precautions must be taken while handling the c.m.o.s. IC3. Damage may result if a static charge is applied to its pins. Check, crop and solder. Switch on. The output of IC8 should be +12V. If it is not, check that R8 has at least -7V on it and that pin 10 of IC3 has half of the voltage at R8. If not, then the i.c. is suspect. If so, then check the positive end of D1 which should have more than +15V on it. If not, suspect Tr3. Switch off.

Minus 5V supply: Load IC2 and C11. Switch on and measure the output of IC2, it should be -5V. Switch off.

Amplifier: Load R11, D2, D3, R12 to R19 inclusive, C7, C8, IC4 and VR1. Check, crop and solder. Move the coil away from any metal and turn on. Adjust VR1 to give an output voltage of +0.5V on pin 6 of IC4. Move the coil to a large piece of iron and the voltage should go up to 0.7V or so.

If the output cannot be set to 0.5V then check the voltages on pins 2 and 3. They should be nearly the same at 0V. If not, fit a new 709. If they are correct solder a $100k\Omega$ resistor across R15 and check that the 709 output can be adjusted. If not, fit a new 709 but if so, replace R15 with the largest resistor that will allow the output to be adjusted from rail to rail. Set the output to +0.5V and switch off.

Sample Pulse generator: Load C17 to C20 and R31 to R35 inclusive. Ensure that IC7 is a buffered type and insert it carefully. Check, crop and solder. Switch on and try to measure the output of IC7 at pin 10. As the waveform is a square wave of duty cycle $0\cdot 1$ per cent the voltage should be very close to $-10\cdot 5$ V. Switch off.

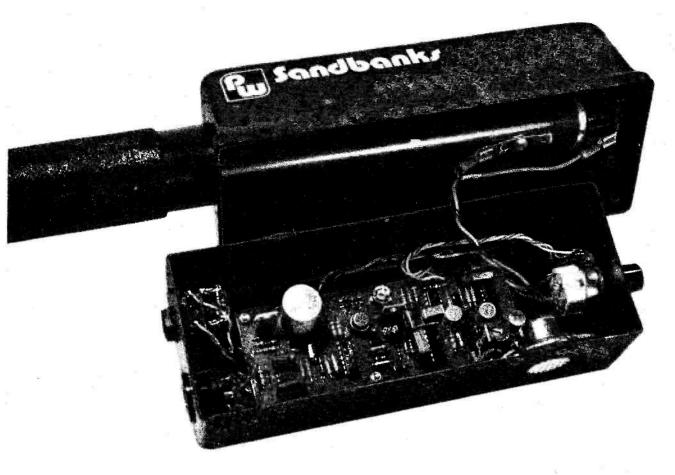
Sample Circuit and Integrator: Load C10, Tr4, R20 to R24 inclusive, R36, C12, IC5 and wire up VR2. Check, crop and solder. Switch on with the meter between pin 6, IC5 and 0V. Move the coil away from metal and set VR2 so that the output is about 0V. Move the

coil near metal and the voltage should rise to almost +12V. If the output cannot be set to 0V measure the input voltages on pins 2 and 3. Pin 2 should be slightly negative and pin 3 should be adjustable to either side of it. If the voltages are wildly out suspect IC7. If they are very near, alter the value of R22 or R23 to bring the range of VR2 to the correct point. Switch off.

Buffer and VCO: Load R25, Tr6, Tr5, R26 to R29 inclusive, C13 to C16 inclusive, IC6 and wire temporarily to the speaker. Check, crop and solder. Switch on with the coil away from any metal. Ensure that with VR2 right down the speaker is silent, and with VR2 right up and the coil near metal, the tone is at its highest pitch. This can be adjusted slightly by altering VR1, keeping the output of IC4 within the limits of 0V to +0.5V.

Final Assembly

When the circuit is working correctly, fix the coil plug PL1, jack JK1 and potentiometer VR2 into the bottom tray, and leave flying leads from the p.c.b. Mount the p.c.b. from 3 screws through the case with suitable spacers, and solder the leads to PL1 and JK1



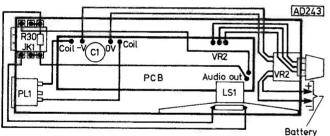


Fig. 7: Wiring details for the case assembly

as shown in Fig. 7. Solder R30 across the jack and glue the speaker into the case with Evostik. Slide the battery tube through the top half of the case and glue the battery cap to the case with Cyanolit. Cement the tube to the inside of the case and the battery cap to the case with ABS cement. Fit the bottom half of the case into position and drill 4 holes for self-tapping screws, avoiding the speaker.

Assemble the complete detector and wind the lead from the search coil around the shaft as shown in the photographs to keep it tidy. Cut the lead to a suitable length, then strip it and connect it to the free socket SK1

Alternative cases

As mentioned previously, this circuit can run from a supply between 9V and 12V without modification, so the battery used is up to the builder. A PP9 will last up to 15 hours, HP2s up to 80 hours and MN1300s up to 200 hours. Should the constructor decide to

build his own case then a visit to the local builders' merchant will secure most of the necessary items for the shaft and handle.

The coil should be wound as detailed but it could be glued to a piece of glass fibre p.c.b. material with all of the copper etched away. The coil must be rigid and waterproof so it is best to cover it with Araldite. The shaft and handle can be made from ³4in waterpipe but the best material in ³4T which has a thick wall suitable for threading. There are a number of proprietary brands of case available to house the electronics. The part of the shaft nearest the coil must be non-metallic.

In use

Whether you use a kit or make your own, the way to use the machine is exactly the same. Best results will be obtained when the user has experience with his machine.

Switch on and turn up the control until the speaker is at its highest pitch. Back the control off until the speaker has just stopped clicking. If the control is set too close to the quiet point then as the coil changes direction through the magnetic field of the earth, it will give an output, so set it back just a fraction more. The best point will be found with lots of practice.

If you are sure there is something to detect and know roughly where it is then the machine can be set with the speaker just clicking at about 1Hz. Search very slowly and listen for a change in the rate of clicking. The machine is at its most sensitive at this point. When you are very familiar with your machine you may wish to alter the value of C14 to 47nF which increases the sensitivity to small objects but makes the control more critical.

Follow-up to **Comparison of the Comparison of

Dick Ganderton

The PW Purbeck oscilloscope probably rates as the most complex project published in this magazine for some time, and as such it is only to be expected that there would be some problem areas to be ironed out.

Although several errors and omissions occurred during the six months the series ran, as far as we know only one 'catastrophic' error occurred and the author has built another model using the kit of parts supplied by Watford Electronics and following the instructions published to the letter. I understand that this instrument failed to work first time due to errors in his work, not in the text. Clearing these faults allowed the scope to work perfectly. This should give renewed heart to constructors who found that on switch-on their scope also failed.

In this article I will try to cover those points which, if readers' letters are anything to go by, have given the most problems.

Y Input Low

By far the largest number of queries received have asked the question, 'where does pin 8 board 3 go to?' Both the author and myself thought that the description 'Y Input Low' would be understood by everyone. Well, we were both proved wrong. The term 'Y Input Low' refers to the low side of the Y signal input. Now the next problem arises with the very high bandwidth of the 'Purbeck'. The scope will look at signals at 21MHz, at a reduced trace height of course, and at this sort of frequency you have to watch out with earthing arrangements. This means that the Y amplifier input low side must be taken right back to the Y Input socket on the front panel to avoid picking up any unwanted currents passing along the other earthing connections.

Ground Planes

This also accounts for the author's use of Ground Planes on the X and Y amplifier boards.

This technique baffled a few readers, but is really quite simple to use as far as construction is concerned. However, it is vital that the wiring layout shown is followed closely. If you choose to ignore it then you will almost certainly have instability problems. It is also vital to ensure that the components are firmly pushed against the copper side of the p.c.b. and that the wire connections are also made cleanly and as close to the board as possible. The use of i.c.

sockets is also taboo, so obviously care will have to be taken when soldering to the i.c. pins.

Power Supplies

After the Y Input Low problem the next most common difficulty was with the power supplies. Here there are several different problems which, if more than one of them occurs in a supply compounds the troubles.

The transformer specified was specially designed for this project by Barrie Electronics and no design details have been given to avoid the possibility of constructors using a transformer with unsuitable characteristics. The way in which the winding resistances are distributed is of vital importance in this instrument and so it was decided that only the Barrie transformer would be specified.

Unfortunately the transformer supplied by Watford Electronics with their kit does not meet the original specification with regard to winding resistances and so if you are using a Watford tranformer you will need to make several modifications to the power supply boards.

If you measure the '250V' output of a Watford transformer on load you will find that it is only about 235V. As this is the input to the tripler circuit it is obvious that the 15V loss becomes 45V at the tripler output. To compensate for this R103 should be replaced by a shorting link and also R101 should be reduced to around 120Ω (still 5W wire wound).

It must be stressed that these changes must only be carried out if you are using a Watford transformer. If you have the correct Barrie transformer then these resistor values must be left as originally specified.

Watford have told us that they will be supplying a correctly specified transformer with future kits and that these will have a 220V extra primary tap, as on the Barrie version. This will probably be the only means of identification between the two types.

Stabilisers

The $-800\mathrm{V}$ stabilised supply proved to be a problem, with several readers not being able to achieve the full output. Assuming that you have 250V going into the tripler then if you cannot get $-800\mathrm{V}$ at pin 7 of board 1 the first thing to check is the voltage across each of the five BZX61C160 zener diodes. If you find that one or more are below the nominal 160V then

you can reduce the value of R104 until -800V is achieved. If the voltage across any of these diodes is zero or very low then the diode is faulty and should

be changed.

The +150V stabilised supply also gives some trouble, especially if you like to test each circuit as you build it. The author designed this regulator to cope with short circuit conditions but forgot the open circuit conditions. If the supplies are tested off load then the +300V raw supply rises to around +370Vand the voltage across R203 rises to around 220V with the result that it can no longer cope with the power and overheats. The answer is to replace R203 with a small wire wound type and replace D201,202 with BZX61C75 types. This will allow the supply to cope with open circuit conditions for a reasonable length of time.

With correct voltages at the outputs of the power supplies we can turn to the more exciting parts of the instrument.

Amplifiers

The two amplifier boards have been discussed earlier in this article in general terms but there are several points to note if problems arise.

If you have any form of instability in the Y amplifier, which will show up as an unwanted trace on the screen, the first thing to check is that you have followed the wiring layout and have not missed out any connections or joined up the wrong pins. If the board is correct then check your earthing system. Note particularly that Pin 8 Board 3 goes direct to the tag under the Y input socket on the front panel. that the third (ground plane) flying lead at the end of the board opposite to the edge connector goes via a short length of wire to a tag under the clamp holding C20 to the chassis and that the paint is cleaned away to bare metal at each earthing point.

The resistors and chokes fitted into the flying leads from the X and Y amps to the tube plates must be mounted as close to the boards as possible and kept separate from each other.

If you still have instability occurring make another

check on all your wiring.

It has been found that in some cases the LM304 regulator oscillates feeding a saw-tooth waveform down the supply rails. The cure is to add a 400pF polystyrene capacitor to Pin 4 of IC304 and ground.

Voltages at various points on the Y amp. are given in Table 1 and these should help with fault finding. The trace should be centred and a $20k\Omega/V$ meter used for measuring the potentials. To avoid instability in the Y amp. a $10k\Omega$ resistor should be inserted in each test lead when measuring the Y output voltages.

The X Amp. seems to have given a lot less trouble than the rest of the scope but Table 2 gives d.c. potentials at various points. As a large part of this board is digital in nature then it is a simple matter to check potentials at the input and outputs of the various gates.

If you have had trouble in getting the trace length down to the specified length of 10 divisions R434 can be increased until it is possible to adjust the trace length with VR408 to just below or above 10 divisions.

The 'catastrophic' error occurred on Board 4 and concerns the omission of a connecting wire from the junction of R424,425,426,427 to the collector of Tr404. Without this wire there is no +12V on the resistorpot. network associated with S401.

Several readers have commented on the slow action

TABLE 1

The state of the s
##NGG 등 10 - 1222 이 경기 2m 이 ##G (All Control All Mark Front Holler Control All Mark Front Holler Co
Y Amplifier d.c. potentials.
IC301 pins 7, 8 +2 3V
IC301 pins 1, 14
Tr305, 308 emitters +1 6V
R330, 331, 333 junction —6.8V
Y_1, Y_2 outputs $+63V$
C306 + 120V
Total current from +150V supply 30mA
(All voltages measured with trace central, all supplies
correctly set and using a 20kΩ/V meter.)

TABLE 2

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١.	X Amplifier d.c. potentials.
	v. Combiners, and Karamanara.
	Education and the control of the state of th
	Edge connector pin 19 +5V (Timebase in Cal. position.)
	Tr411, 410 emitters +4.5V
	Tr408, 409 emitters + 0.4V
	Tr408, 409 emitters + 0.4V
	n v_iii vita ta t
h.,	Tr407 emitter -6.8V
	X ₁ , X ₂ outputs +78V
: .	X ₁ , X ₂ outputs +78V
	in 171 - 27 chaile for the 181 - Charles and Label 2 in the first of the first first of the first terms where the
	Total current from +150V supply 20mA
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of the brilliance control and the slight drift in the vertical position of the trace after operating the vertical shift control. The first effect is normal and one just learns to live with it. The second is caused by thermal changes in the Y Amp. output stage transistors. Increasing the size of heat sink on these and adding heat sinks to Tr305,306,307 and 308 will help to reduce the drift. By the way all these transistors run very warm in normal use. If the flyback trace is visible it is more than likely that you have the brilliance too high.

Front Panel

A couple of minor errors crept into the front panel wiring drawing in the third part of the series. Section B of S4 should go to Board 4 Pin 18. The Trig. level pot. VR1 is connected between +12V stab. and -12Vstab., the annotation slipped down slightly on the drawing board. VR4 should be connected to -6V via a 680Ω resistor (instead of ground), to give equal swings on Ext. X in.

The case work should prove quite straightforward as those readers who have used the Bazelli case will have most of the work already done while it is probably fair to assume that those who decide to make their own case will at least know how. For those who do make their own case the transparent overlay is still available from the editorial offices at £2.25 including p&p. This overlay also has a fine photographically reproduced graticule, available separately at 60p including p&p, and has been in service now for over nine months on the author's scope without any ill effects.

No drawings were given for the drilling of fixing holes in the back and bottom panels as these are marked using the boards and components themselves. Remember to clean back the paint at all the earthing points.

Brackets

The small brackets shown in the last part are simple to make. A couple of errors crept in on brackets (a) and (c). In bracket (a) the length of the Continued on page 58

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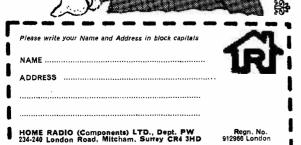
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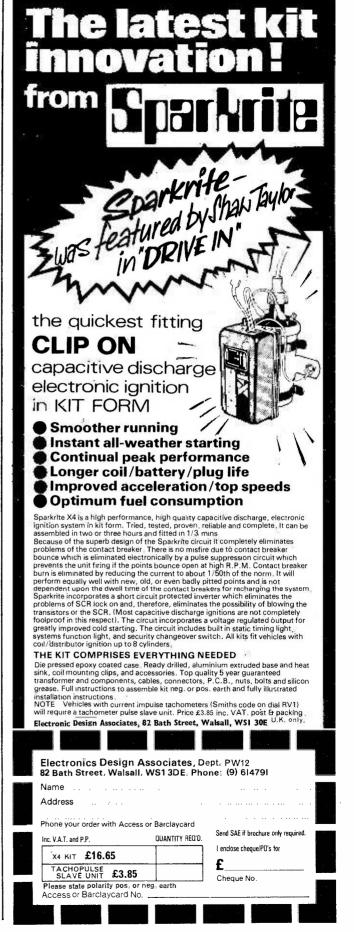
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PW Purbeck Follow-up

Continued from page 56

longer arm should be 100mm while in (c) the distance between the two large holes should be 38mm at 16mm from the bottom edge.

The mu-metal shield around the tube neck is not connected to ground—or earth or 0V, as it is only needed for magnetic shielding.

The PW Purbeck offers the home constructor the opportunity to build and use a scope with a performance which is comparable to many commercial instruments selling at a much higher price. With the capability to display waveforms, at reduced amplitude, up to 21MHz the constructor has an oscilloscope which should prove to be very useful.

In conclusion I would like to thank all those readers who have written to me or the author with their problems. This feedback has helped us to sort out most of the problems as well as showing us that this pro-

ject has been very well received.

For readers who have access to another oscilloscope a sheet showing waveforms at various points of the X Amplifier board is available, on receipt of an s.a.e. from the editorial offices, Practical Wireless, IPC Magazines, Westover House, West Quay Road, Poole, Dorset.



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F.M. RECEIVERS

DEVICES & CIRCUITS

PART 1

M. J. DARBY

Well-designed frequency modulation receivers can provide high-quality reception, with a lower noise component than that obtained with amplitude modulation. Although narrow bandwidth f.m. transmissions are used for special purposes, good quality f.m. broadcasting demands a wide bandwidth, typically 300kHz, and this cannot be practically accommodated unless a very high frequency carrier wave is employed. An internationally agreed band between 88 and 108MHz has therefore been allocated for the purpose.

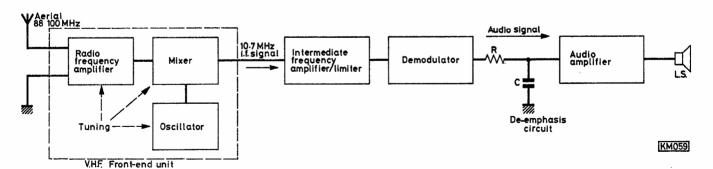
General Propagation Conditions

Transmissions at v.h.f. suffer little reflection in the upper atmosphere and are therefore normally only received at places of virtual line-of-sight distance from the transmitter. This factor may be something of a frustration to the DX enthusiast, but offers the advantage of being reasonably free from interference by other stations. Only under exceptional circumstances of atmospheric disturbance does the v.h.f. signal travel longer paths, and these "openings" will be readily apparent when they do occur.

Noise Levels

In the case of v.h.f. stereophony, using f.m. techniques, a greater signal bandwidth is required than would be the norm for an equivalent monophonic channel. However, the wider bandwidth necessary for stereophony results in an inferior signal-to-noise ratio, the degradation being in the order of 20dB. Consequently, if one is not prepared to install a quality aerial of high gain or if one is unfortunate enough to live in a particularly difficult area, then the choice may have to be made between a noisy stereo signal or more reasonable monophonic reception.

A further reduction in the noise levels of f.m. receivers is achieved by the use of h.f. pre-emphasis at the transmitter and subsequent de-emphasis within the receiver. The higher audio frequencies are therefore transmitted at an increased amplitude and reprocessed when reaching the receiver until the overall response is level at all frequencies. Any noise in the aerial or early stages of the receiver will take the form of a hiss, and the majority of this energy will therefore be at high audio frequency. This is



Frequency modulation has many other bonuses to offer, one of which is that it enables a considerable reduction in noise level to be obtained. Noise generated by vehicle ignition systems for example, is to a large extent amplitude modulated and receivers may be so designed to afford a high level of rejection, typically 45 to 65dB, to the a.m. signal.

Although the technology to broadcast stereophony on medium frequencies using a.m. (PW, March 1978, page 813; April 1978, page 891) has been around for some time, at present the sole source of regular stereophonic broadcasts is within Band 2, 88-108MHz and is invariably f.m.

Fig. 1: Block diagram of a full monophonic f.m. receiver

ultimately attenuated by the resistance/capacitance filter which forms the de-emphasis circuitry within the receiver.

Certain overseas stations use proprietary noise-reduction systems, such as that developed by Ray Dolby. Some tests with Dolby were conducted in the UK by the IBA a short time ago, but as yet no definite steps towards adopting the system appear to have been taken. Many of the small radio stations in the United States have opted to use Dolby, however.

Basic Receiver

In a typical f.m. receiver, the v.h.f. signal is converted to a conventional i.f. of 10·7MHz. All high-quality receivers employ superheterodyne principles due to the difficulties involved in direct amplification at these frequencies and the problem of obtaining an acceptable degree of selectivity.

The basic circuit of an f.m. monophonic receiver is given in Fig. 1. A front-end unit amplifies the in-

Stereo

A block diagram of a stereo f.m. receiver is shown in Fig. 2. Note that, unlike the monaural receiver of Fig. 1, no audio filter for de-emphasis follows the demodulator circuit. Two separate audio de-emphasis filters are required after the stereo decoder, the component values being calculated as indicated previously.

We will now look at the various parts of typical

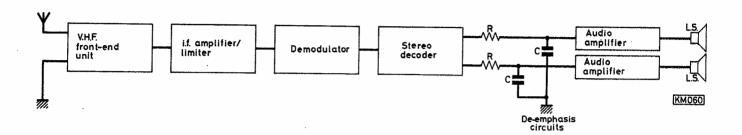


Fig. 2: A full stereo receiver block diagram

coming signal and converts it to the 10.7MHz intermediate frequency. Importance is given to the fact that the incoming signal should be amplified before the i.f. stages. Frequency changing mixer circuits are inclined to add more noise than a good amplifying stage. If this technique is employed, the signal will normally be appreciably larger than the noise added by the mixer.

The three stages of the front end must be tuned in such a way that the difference between the incoming signal, irrespective of its frequency, and the oscillator is $10.7 \mathrm{MHz}$. This, of course, is the intermediate frequency.

Tuning can be carried out either with a conventional air-spaced tuning capacitor or by a variable-capacitance diode ("Varicap"). In the latter instance, frequency variation is achieved by altering the diode capacitance. This is most usually done by slight adjustment of the voltage via a potentiometer.

From the front end, the 10.7MHz signal is passed to the amplifier/limiter, which consists of circuitry designed to tailor the response to the correct characteristics. Early receivers used tuned circuits for this purpose, but the modern approach is to employ ceramic filters, which require no alignment and provide good selectivity and frequency response.

The limiter circuit controls the amplitude of its output within a constant positive and negative peak range. All the signals from the limiter therefore have the same amplitude and this greatly reduces the effects of any amplitude modulation which may also be present as a spurious signal.

The limited signal is then demodulated; this means it is converted into an audio signal. The latter is passed through a simple de-emphasis filter connected as shown in Fig. 1 and hence to an audio amplifier and loudspeaker. In Europe the de-emphasis time constant is 50μ s, whereas in the USA it is 75μ s. This means that in Europe the product of R and C of the filter in Fig. 1 must be 50μ s or thereabouts, for example, R may be $12k\Omega$ and C 3.9nF or alternatively R may be $4.7k\Omega$ and C 10nF.

f.m. receivers in more detail, but only modern circuits will be covered using integrated circuits for simplicity.

The front end

The components in the front end of an f.m. receiver operate at frequencies of the order of 100MHz. At such frequencies the placing of components and the lead lengths can be very critical, so it is not especially easy for the home constructor to build a front end using either discrete components or one of the relatively few integrated circuits which have been made available for this application.

Commercially manufactured front end units are readily available, completely enclosed in their screening metal, providing the constructor with a simple way of avoiding problems associated with home construction, such as radiation from front end to a.f. amplifier, etc. The use of a commercially made front end unit makes the construction of an f.m. receiver somewhat easier than the construction of an a.m. receiver. Both ganged-capacitor tuned types and varicap tuned types are available, but the latter are now more commonly used by constructors. However, the

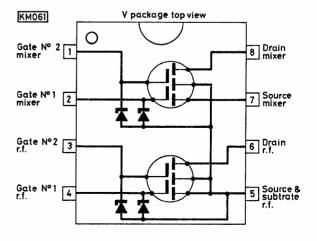


Fig. 3: Connections and internal circuit of the SD6000

use of a varicap tuned front-end involves slightly more complex circuitry, since a regulated voltage supply is essential for the tuning system. The tuning control in varicap tuned receivers is a multi-turn helical potentiometer (such as the Beckman Instruments 10-turn miniature type 7276) or a long linear potentiometer. The use of such varicap tuning systems reduces the mechanical problems associated with ganged capacitor tuned front-ends which must be provided with some form of a pointer and with a no-backlash slow motion drive. A varicap front-end can be placed where convenient, but a ganged capacitor tuned front-end must be placed in such a position that the control is conveniently on the front tuning panel.

An automatic gain control voltage may be fed through R1 to the second gate of the r.f. stage at pin 3. As shown in Fig. 5, this voltage can provide a range of gain of 50dB, the gain decreasing as the bias falls from about +8V to 0V.

The output signal from the r.f. stage is developed across the choke L4 and is coupled through C6 to the r.f. tuned circuit comprising L2, C1b, C7 and the small trimming capacitor. It is then coupled by C8 into pin 2 of the SD6000 which is the input gate of the mixer section; R4 and R5 provide a suitable bias voltage to this gate.

A 2N4126 pnp transistor is used as an oscillator stage, this stage being tuned by L3, C1c and the trimmer C13. The oscillator voltage is coupled by C12

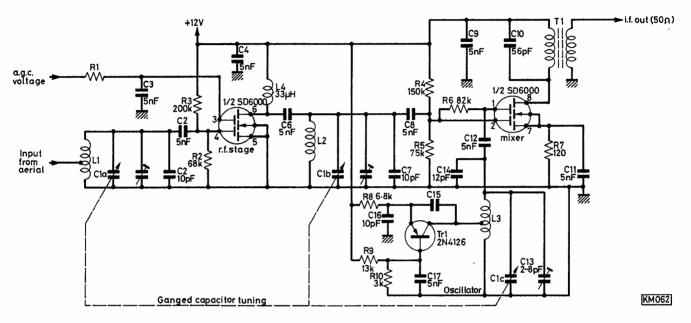


Fig. 4: Typical front end circuit using capacitor tuning

into gate 2 (pin 1) of the SD6000 mixer stage. The output from the mixer appears across the primary of T1 which is resonant at 10 7MHz. The secondary

Front end Chips

At the present time only two integrated circuits are readily available for use in front-ends. These are the Signetics SD6000 and the AEG-Telefunken type TDA 1062.

The SD6000 device is encapsulated in an 8-pin dual-in-line plastic package with the connections shown in Fig. 3. It contains two Diffused Metal Oxide Silicon (D-MOS) field effect transistors for use as the signal frequency amplifier and mixer. In addition, this device contains four Zener diodes one of which is connected in the circuit of each gate electrode to protect the D-MOS devices against breakdown by the accumulation of electrostatic charge. If stray charges produce more than a certain voltage across one of the gate protecting Zeners, the Zener conducts the charge to the substrate and thus protects the D-MOS device against breakdown of the thin silicon dioxide layer by an excessive voltage.

A ganged capacitor tuned circuit using the SD6000 device is shown in Fig. 4. The signal from the aerial is fed to the aerial tuned circuit comprising L1, C1a and a small trimmer in parallel with C2. This signal is applied through a 5nF capacitor to the gate of the r.f. amplifier at pin 4 of the SD6000; a suitable bias is also applied to this device by means of the voltage divider of R2 and R3.

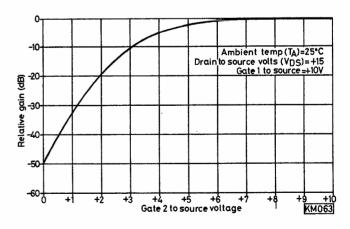


Fig. 5: Link between gain and bias level in the SD6000

of this transformer is a 50 ohm coupling winding which provides the output to the I.F. stages.

The SD6000 circuit shown can provide very low noise (noise factor about 3dB at 100MHz) and a power gain of some 30dB or more. The use of the second gate of the mixer for the oscillator voltage enables very high isolation of this oscillator frequency from

the input to be obtained and prevents frequency pulling of the oscillator signal.

The TDA 1062

The TDA 1062 front-end device can be used at frequencies of up to 200MHz and is available from several suppliers. This monolithic device includes an r.f. amplifier, mixer, oscillator, a.g.c. amplifier and a voltage stabiliser on one chip, so represents a complete v.h.f. front-end.

The device itself requires a supply voltage of between 9 and 15 volts at a typical current of 30mA. A particular feature of the circuit shown in Fig. 6 is the use of BB104 double varicap diodes for tuning without any parallel trimming capacitors. This keeps the minimum capacitance very low and allows the frequency range of 88 to 108MHz to be covered with a tuning voltage of only 2V (minimum) to 7.5V (maximum).

coil L3 and the BB104 back-to-back diodes marked D2. The signal is magnetically coupled from L3 to L1, the centres of these coils being 1.3cm apart. The tuned circuit of L1 and D3 is coupled by L2 to the mixer input at pin 4.

The output from the mixer passes through an internal low-pass filter to the output at pins 13 and 14 and hence to the 10·7MHz output tuned circuit of L8 and L9. The oscillator circuit is connected by the coupling coil L6 to pins 1 and 16. The oscillator tuned circuit itself consists of L5 and the BB104 tuning diodes D4. The TDA 1062 contains an internal voltage stabiliser circuit which powers the oscillator and mixer and greatly assists thermal stability.

The output signal from pin 13 is also used to provide automatic gain control. This signal is coupled through the 10pF capacitor to the junction of the two 1N914 diodes which rectify it and develop a positive voltage across the 4.7nF capacitor which is proportional to the i.f. output signal. This positive voltage

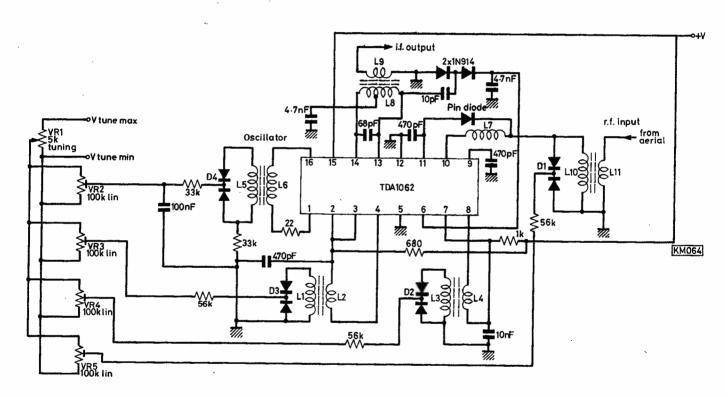


Fig. 6: A complete varicap-tuned receiver using the TDA 1062: max. tuning voltage should be 7.5V and min. 2V (regulated). Trimmers VR2-5 are adjusted for optimum tracking

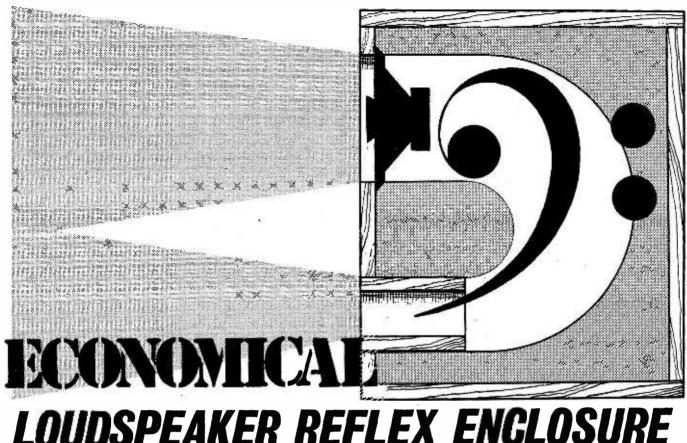
The circuit of Fig. 6 provides a power gain of typically 30dB. The typical noise figure of 5.5dB is not quite as good as that quoted for the SD6000 circuit, but is better than that of many commercially produced front-end modules. The signal frequency bandwidth is about 1.7MHz and the intermediate frequency bandwidth about 500kHz. The three signal frequency tuned circuits provide an image rejection of some 80dB, whilst the i.f. rejection is about 100dB.

The signal frequency input is coupled by L11 to the tuned circuit consisting of L10 and the BB104 diodes marked D1. The signal then passes through L7 to the radio frequency amplifier input at pin 10. The output from this amplifier appears at pin 8 and is coupled by L4 to the tuned circuit comprising the

is fed to the input of an a.g.c. amplifier at pin 6 and the output of this amplifier at pin 11 is used to feed the p-i-n diode which provides the required a.g.c. voltage to the input at pin 10.

Next Instalment

This article has considered general points about f.m. receivers and has covered some front-end circuits using i.c.s. The output from the front-end unit (no matter whether it is a commercially manufactured front-end or a home constructed unit) must be fed into a suitable intermediate frequency amplifier. The next article will deal with various i.f. amplifier circuits.



LOUDSPEAKER REFLEX ENGLOSU E.M.Parratt BA

Anyone with an interest in good quality audio is likely to be interested in efficiency, especially at the "business" end (the speakers) which represents the final interface with the listener. Much has been written about the advantages of particular speaker and enclosure types, and it is now common knowledge that the infinite (closed) baffle needs the greatest drive, the reflex much less, and the horn hardly any at all, but designs utilising a redundant fireplace as the exit of an underfloor horn leave a lot to be desired in terms of portability.

Choice of System

The bass reflex, properly cut, can provide a good compromise—sensitivity (efficiency) is quite high, and construction is pretty simple. The conflicting problems of dynamic range, efficiency, enclosure resonance, and the extension of bass are probably best solved by the acoustic labyrinth, or transmission line enclosure (basically a tuned resonator), but this is highly complex in construction and involves the use of several speakers in one cabinet. Further, crossover units are required, with resultant phase-shift and additional expense.

It's quite true that floor-standing reflex enclosures can sometimes present a problem in matching to the room characteristics, but this design (which is also eminently suited to the "Wimborne" Music Centre) is sufficiently compact to mount as shelf speakers, although the heavy timber construction will require firm shelving.

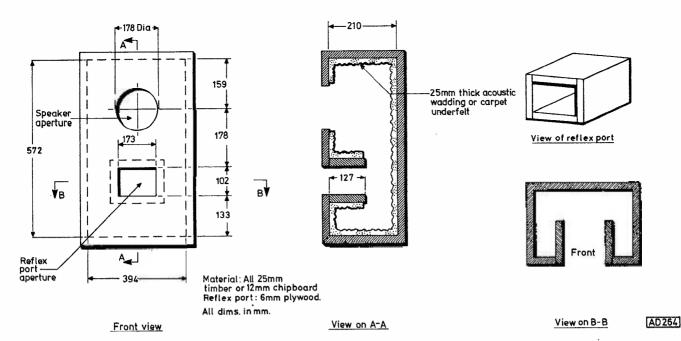
Specification of Prototype Enclosure

System type; bass reflex ("loaded port") Enclosure volume: 50 litres (3000 cu. in) each unit Overall dimensions: height 622mm, width 444mm, Drive unit: Elsc BNC298--impedance 80, twinpaper cone, power handling 15W r.m.s. Frequency range: 40Hz-18kHz - 3dB.

Comparisons

A "learned" article recently appeared in one of the Sunday newspapers concerning the terrors of a new Japanese assault upon the "British sound" in speaker units. It seems that £350 is regarded as a reasonable price to pay for a good Japanese pair! Whatever the advantages of such a set-up (presumably solid gold speech coil dome?), the design described here will give a reasonable performance over the range 40Hz to $18kHz \pm 3dB$ for about £25 the pair. The units are suitable for any amplifier with outputs at 8Ω and r.m.s. output levels up to about 20W.

One reason for choosing the reflex technique is that the alternative simple form (the infinite baffle) requires a much larger enclosure volume for adequate bass extension, and since the intention is to move quantities of air efficiently, the reflex type is the obvious choice.



General constructional details of the reflex closure

The Drive Unit

The Elac 8NC298 8in twin-cone unit was chosen for several reasons: (a) it's cheap, (b) it offers a wide frequency range, (c) the tweeter cone obviates the need for a separate pressure unit or dome tweeter. and (d) its relatively small cone combined with the favourable loading of the port results in a fair level of efficiency. In terms of initial cost, it seems a good unit anyway, featuring a ceramic magnet (flux density 12,000 gauss), rolled suspension surround maintaining excellent stability at low frequencies, and a specially stiffened tweeter cone. A further advantage of the twin cone approach is that there is no chance of damaging the speech coil of a typical tweeter unit by instantaneous peak h.f. levels. This is a more common occurrence than one might suppose from its lack of publicity, and the problem has increased since the advent of direct-coupled audio power stages, in some cases the amplitude of a "fast-wound" tape signal being sufficient to destroy a fairly expensive h.f. unit.

The speaker is available from Wilmslow Audio, (Dept. PW), Swan Works, Banks Square, Wilmslow, Cheshire SK9 1HF at $\pounds 6 \cdot 75 + 75p$ postage and packing.

Enclosure Construction

Each box is constructed from 25mm or 18mm timber or chipboard. The idea is to obtain a rigid frame so as to minimise colouration from cabinet resonance, but corner pieces *must not* be used or the internal volume will be altered, with loss of bass as the audible result of incorrect phasing with the reflex port. Glued dove-tailing is the preferred method of construction, although more troublesome of course, Evo-Stik "Resin W" or "Cascamite" are suitable wood glues.

All the inside surfaces (except for a 20mm gap around the speaker aperture) are lined with acoustic wadding or carpet underfelt to reduce reflections. The apertures may be covered with light speaker fret or left unprotected according to inclination.

Connections to the speaker unit itself are made before fitting the top panel, and a suitable hole or slot cut to take the cable through the wood. This is then packed with glue around the cable in order to obtain a good airtight seal throughout.

LETTERS

Sir: I was rather disappointed at the rather negative article on Norton CDAs in the October 1978 issue of PW. It is true that these devices are inferior in most respects to 741s say, but what most people don't appreciate is that for fifty per cent of all op. amp. applications the 741 is over-specified! For instance, the degree of accuracy in the definition of closed-loop gain that results from a very high open-loop gain is often unnecessary. If only a little more effort were made to sell the idea of CDAs to the public, the rather inflated prices being charged by most retailers would soon drop. Some retailers have in fact been offering LM3900s at 50p or less for quite some time. At these sort of prices, (about 12p per amp) their true role becomes clear. They are cheap substitutes for 741s in non-critical applications. In the USA this is more fully appreciated. The CDA is more often incorporated, where appropriate, in published circuit designs and the CDA/741 price ratio is about two to one, not three or four to one as over here.

The British hobbyist is as quick as any to adopt the latest Large-Scale-Do-It-All-For-You-i.c.s, but rather slow to adapt to new technology, discrete components and general-purpose i.c.s, requiring new approaches to circuit design. Magazines can help with positive articles like J. B. Dance's on VMOS. Retailers can help by anticipating popular interest in new technologies and keeping prices low. Between them they can help start the downward spiral in cost that results when popularity and cheapness feed on each other.

P. J. Rimmer London N4

PRODUCTION LINES alan martin

Mains intercom

A new f.m. Wireless Intercom Model No. LP1010 introduced by Hadley provides a three channel communication system.

No wiring between the stations is required, as connection is made when each unit is plugged into the mains.

Having three separate selective channels, the LP1010 caters for a wide variety of communication needs without the inconvenience or cost of installation work.

A system of three stations costs £120 plus VAT, additional stations cost £40 each plus VAT. Further information from: Hadley Sales Services, 112 Gilbert Road, Smethwick, Warley, West Midlands B66 4PZ. Tel: 021-558 3585.

Mini i.f.

Available from Ambit, the 5S i.f. transformer, probably the world's smallest fully tunable i.f. transformer for frequencies from 100kHz to 15MHz.

The transformers are based on values scaled down from the existing standard ranges in 10mm and 7mm format i.f. transformers.

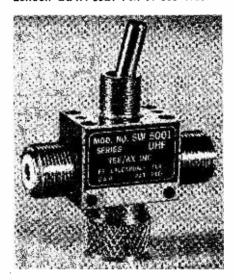
The applications of such a miniature coil are many, ranging from ultra slim broadcast and paging receivers, to miniature filter blocks, transceivers, t.v. and instrumentation applications where tunable inductors are required.

Costing 44p plus 12½% VAT and 25p P&P, the transformers and specification sheet are obtainable from: Ambit International, 2 Gresham Road, Brentwood, Essex. Tel: (0277) 227050.



Coaxial switches and relays

An interesting new range of coaxial switches and relays is available from the Tee/Ax Corporation of Florida. The toggle switches are also obtainable in multiple banks or ganged. Further details from the UK distributors, IMOS Ltd., Suite 307, Shoe Lane, London EC4A 3JB, Tel. 01-353 4133.



Mixer kit

Recently announced the introduction of the Prokit 62 mixer. The six input, stereo output unit features fully professional specifications offering line and mic. inputs, treble and bass equalisation, echo and cue busses. pan and long throw fader on each channel. A unique feature of this mixer is that it is available only in kit form. There is a 32 page manual for assembly, which covers all areas from soldering to component indentification and insertion, to, the full test and fault finding schedule. The simple to follow instructions make construction a couple of evenings work, even for a relative beginner. The unit costs £99.95 which includes VAT, a +15V, -15V power supply is also available at £15.00 which includes VAT, post and packing for either or both units is £1.50. Further details from: Turnkey, 8 East Barnet Road, New Barnet, Herts. Tel: 01-440 9221.

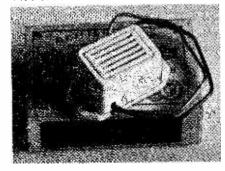
Mini-Buzz

FieldTech Ltd., is able to offer a range of miniature d.c. solid state electronic buzzers, which utilize transistor oscillators and have no mechanical points to arc or require maintenance.

Six variations are available operating on 1·5V, 3V, 6V, 9V, 12V and 24V d.c. and each weighs only 10 grams. The 1·5V and 3V types have a 65dB min. output at 1m, whilst the other four have a 70dB min. output at 1m. Being solid state there is no r.f. field generated, and all types provide high reliability with minimum current dissipation.

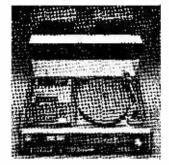
A variety of applications include automotive warning and monitoring indicators, portable and battery operated equipment, test apparatus, timers, intercom and telephone sets, alarm devices, digital clock alarms, etc.

Further details and prices of this low cost range of miniature buzzers are available from: FieldTech Ltd., Components Division, Heathrow Airport-London, Hounslow, Middlesex TW6 3AF. Tel: 01-759 2811.



Practical Wireless, January 1979

PROFESSIONAL KITS THAT SAVE YOU MONEY!!



M C020

AN ADVANCED MUSIC CENTRE for the experienced constructor. This unit is available as a fully wired chassis, in modular form or as a kit. It can be built in easy stages or as a complete unit. A variety of cassette decks are suitable.

SPECIFICATIONS

	SPECIFIC	AIIURS	
AMPLIFIER Power Output: Distortion:	Nominal 2 × 25 watts RMS. THD @ 2 × 20 watts 0·7%		: 475KHz 10∙7KHz
Frequency Range:	@1.5dB 30Hz- 15KHz	AM (internal) AM (external)	Ferrite Rod 2 pin DIN
Tone Control Range: VC~20dB	18dB	FM (external)	Co-axial 75 ohm unbalanced!
Basic Electrical centre Treble Electrical centre	@ 100Hz-14dB @ 10KHz+8dB —14dB	AGC: For 6dB audio change	46dB
Loudness Control: VC30dB	@ 100Hz+14dB		± 1 · 5KHz @ 6dB
Filters:	@ 10KHz+11dB	RF Sensitivity: @ 20dB S/N Ratio 200KHz	1500µV/m
VC30dB Controls:	@ 10KHz—6dB 5 rotary: volume,	600KHz	500μV/m 200μV/m
Controls:	balance, bass, treble	FM:	
Switches:	9 push button: phono, tape, radio, aux. input, mono/ stereo, loudness, fil-	RF Sensitivity @ 26dB S/N (mono) 88MHz 100MHz	2 ·5μ V
	ter, speaker switch- ing, separate mains switch.	@ 46dB S/N (mono) @ 46dB S/N (stereo) Distortion:	16μV 125μV
Meters:	2 signal strength: FM tuning		0.9%
Sockets:	Headphones: 5 pin DIN Aux: AM aerial (ext) FM aerial (co- axial): 4 × 2 pin DIN L/S	企士 1·5dB Stereo Separation: Audio Filter	30Hz-15KHz 40dB Flat to 55KHz 50dB @ 130KHz
SUITABLE SURPLUS	5 MODULES. Fully 1	Wired	





TU020

A Hi Fi tuner amplifier

This unit can be built from our modules or as a complete kit. Input for mag cartridge, tape record/playback, MW/LW/VHF stereo, tuner. Uses the same R F Board as does the Wimborne with birdie filters, multiplex filter, varicap tuning on MW and LW.

Items from the Wimborne numbers 2 and 3 can be used for different

performance specifications.

SPECIFICATIONS

£4.50

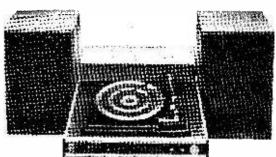
Power output 25 Watts RMS per channel (both channels driven)
Total harmonic distortion 0·05%
Bass 100 Hz ± 12 dB
Treble 10 KHz ± 12 dB
Frequency response ± 1·5 dB 30 Hz—20 KHz

Fully wired modules, Preamplifier, £6.99 £9.50 Power amplifier

RF Board Power supply unit Transfomer

FM sensitivity 1·0 µV for 26 dB S/N ratio IF rejection 60 dB image rejection 60 dB Stereo separation 40 dB AM sensitivity 200 µV at 1600 KHz 20 dB S/N

Magnetic PU amp £7·99 Hardware kit PLEASE ADD £1 for postage and packaging for each item except the mag PU amp which £33.95 £3.99



AU10

A Radio Record Player Kit which has everything you need to make a first class three band STEREO unit. Can be assembled in modular form or from scratch. A professional finish is guaranteed.

2 × 10W RMS, both channels driven

SPECIFICATIONS

Amplifier Output Distortion: Controls:

1% ± 2 × 5 watts

Four rotary 1, OFF/ON/VOLUME 2. Balance 3, Treble

4. Bass Bass @ 100 Hz ±9dB Tone Control Range: Treble @ 10 kHz ±9dB

Outputs: 2 × 2-pin DIN for 8 ohm loudspeakers 1 switched stereo headphone socket 1 5-pin DIN Aux. Tape in/out socket Frequency response @ ± 1.5dB 30 Hz-15kHz Stereo Performance:

Separation 40dB Audio Filter—flat to 55 kHz—50dB @ 130 kHz 7 Push Button: Phono, Tape, Mono, FM, MW, Controls:

LW, AFC 1 Rotary: Tuning

Radio Tuner Medium Wave 525—1620 kHz Long Wave 155—280 kHz FM VHF 88—108 kHz Waveband Coverage: Long Wave FM VHF

8.90 RF Board Price Hardware Kit 13.95 Record Deck 8.75 Amplifier & Pots 6.00 **Dust** cover PSU & Transformer



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The most versatile music centre ever

Complete kit price approximately £110 comparable price £180.

Hardware Kit, for a professional finish, trim, knobs, extrusions, sockets, top moulding 50 plus parts. £9.95.
 Amplifier Module (11-14) watts RMS per channel, wired and tested,

£19 .95, kit price £13 .95.

3 Mechanically tuned RF Board, absolutely complete and tested $1.5\mu V$ sensitivity FET front end 7 way switch, MW, LW, stereo VHF, only requires 13 VDC. £21.95.

ALC tape counter, excellent specification, incredible price £24.95.

Varicap tuned RF Board, the ultimate AM/FM Receiver kit, MOSFET

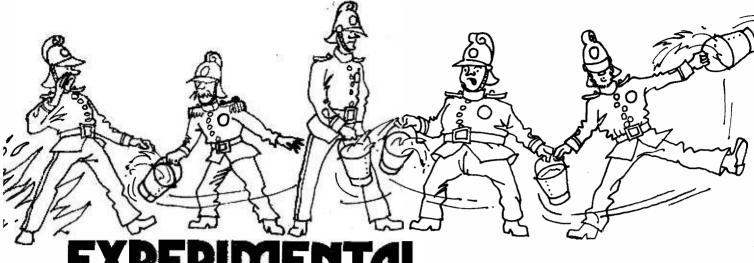
5 Varicap tuned RF Board, the ultimate AM/FM Receiver Kit, MOSFE! front end, Deviation mute, interstation mute, fantastic performance 1.0 µV sensitivity, Hitachi, AM receiver Ic, excellent MW/LW performance, complete with 7 way switch, The basis of the finest Hi Fi system, fully wired and tested. £33.95. Kit £26.95.
6 Magnetic PU Amp with LM387 or discrete (BC 149). Both £2.99 each.
7 Wooden Plinth plus base board £9.95.

8 Perspex dust cover plus hinges £6.50.

9 Mains transformer £2.99. All items include VAT. Please add £1 per item for postage and packaging, item 6 30p. For further details please send 15p.

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EXPERIMENTAL ACOUSTIC DELAY LINE

M. J. HUGHES MA, C. Eng, MIERE

The advent of the "Bucket Brigade" device has opened up a whole new range of prospects to the audio experimenter. It is the possibility of generating fixed, or variable, time delays to audio signals without recourse to electromechanical components—such as tape decks or reverberating springs—that makes these new components so exciting.

Applications

This project provides the home experimenter with a versatile delay unit which can be used to create phasing effects (that strange "whooshing" sound) that are so often used in modern music or, with a change in value of one or two components, it can be used to provide a simple chorus effect and reverberation or echo. A power supply has not been included because most users will want to build such a unit into an existing piece of equipment. To get the system operating you will have to supply +12V and -12V rails but at very low current (about 12 and 8mA respectively) and therefore batteries can be used quite effectively if you just wish to experiment.

The unit has been designed to have virtually zero "insertion loss or gain" which means to say you can insert the system into an audio line without it affecting the matching that already exists between units. With its controls set to minimum effect, any sound will pass through it totally unaffected by its presence. The only proviso is that input signals should not exceed 300mV otherwise distortion will occur. It is therefore ideally suited to go into a phono line between a pre-amplifier and a power amplifier.

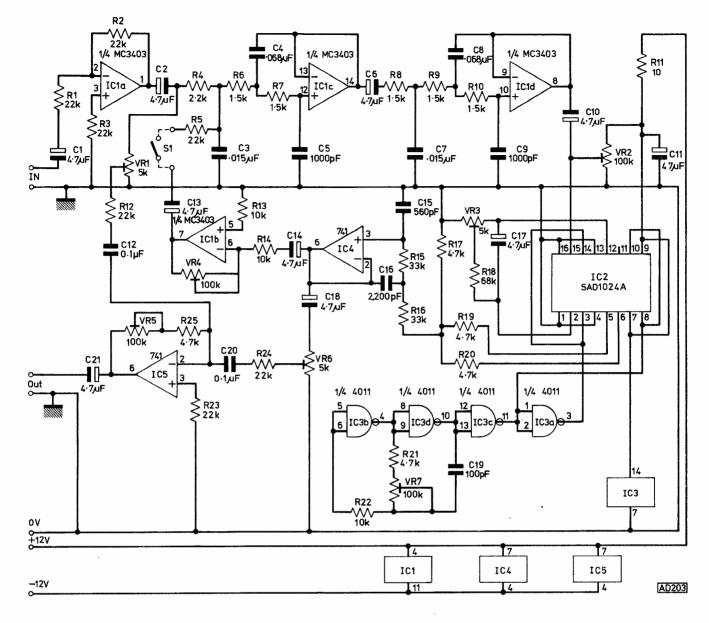
The component values shown in the circuit of Fig. 1 have been selected to give delay times in the range of one to about thirty milliseconds which are best suited to phasing effects. Let us briefly go through the circuit to explain the functions of the various sections.

Circuit Description

The input buffer stage based on IC1a has a medium input impedance and unity gain set by R1 and R2. Insertion gain can be increased by increasing the value of R2. IClc and ICld provide a 36dB per octave top-cut filter to limit input frequencies which are likely to clash with the internal clock frequency of the unit. IC2 is the SAD1024A bucket brigade device which contains two elements each having 512 stage delays—each stage delay is equivalent to half the period of the clock. We are using the two elements in cascade therefore we have 1,024 stages. The input signal to the SAD1024 has to be biased on a pedestal which is set by VR2 for the first element and by VR3 for the second cascaded element. The simple oscillator based on IC3 operates in the region of 50 to 100kHz; its frequency can be controlled by VR7. This oscillator provides the clock for the bucket brigade device, so the delay time can be adjusted with VR7.

The delayed output signal from IC2 has, superimposed upon it, switching "glitches" from the clock together with harmonics. These have to be removed to prevent irritating noise and distortion so the signal is passed through another top-cut filter in the form of IC4. The signal branches at the output of IC4. Part of it is fed to volume control VR6 which adjusts the level of the delayed signal and mixes it with the untreated signal fed to the mixing point via VR1. The mixed signal is then fed to IC5 which acts as an output buffer, the gain of which can be preset by VR5 so that the overall internal losses of the unit are just compensated for.

The other part of the signal from IC4 is fed through a gain-adjustable buffer IC1b and via a switch back to the input of the first top-cut filter. With S1 closed a selectable level of the delayed signal can be fed back and allowed to go through the delay over and over again. If the overall loop gain of this process is greater than unity, "howl round" will occur



—similar to acoustic feedback in a room—therefore VR4 has to be adjusted to prevent this critical situation occurring. The object of feeding back the delayed signal is to increase the reverberation effect.

Fig. 1 : The complete circuit diagram of the Experimental Acoustic Delay Line. R5 should be 2-2k Ω

* components

esistors	Capacitors	Integrated Circuits 741 (8-pin d.i.l.) 2 IC4, 5
₩ 10% - The Think I have the	Plate Ceramic 63V	4011 1 IC3
10Q 1 R11	100pF 1 C19	MC3408 / IC1
1 5kΩ 5 R6, 7, 8, 9, 10	560pF 1 C15	SADIO24A 1 IC2
2·2kΩ 2 R4,5		
4-7kQ 5 R17, 19, 20, 21, 25	Polystyrene	(Herbert Controls Ltd.,
10kΩ 3 R13, 14, 22	1000pF 2 C5,9	Spring Road, Letchworth)
22kΩ 6 R1, 2, 3, 12, 23, 24	2200pF 1 C16	
33kΩ 2 R15, 16		Miscellaneous
68kQ 1 R18	Polyester 250V	S1. s.p.s.t. switch.
	0-015µF 2 C8,7	Printed circuit board
	0.068µF 2 C4.8	
	0·1μF 2 C12,20	
otentiometers		
Miniature horizontal presets	Electrolytic	
5kΩ 3 VR1, 3,6	4·7μF 40V 9 C1, 2, 6, 10, 13, 14, 17,	18,21
100kΩ 4 VR2, 4, 5, 7	47./F 25V 1 C11	

Construction

All components are mounted on a single p.c.b., the track pattern of which is shown in Fig. 2. The component layout is shown in Fig. 3. We have used preset potentiometers for all controls and have deliberately laid out the board in this way to provide the experimental user with a convenient breadboard. The controls which would normally be considered to be front panel controls are VR1, VR4, VR6 and VR7. These have been positioned near the edge of the board so that you can substitute board pins and take flying leads to external controls.

Testing

When the unit is assembled and ready for test you should roughly set all potentiometers as follows:

VR1 — to the ground rail end VR2 — approximately mid way VR3 — approximately mid way

VR4 — to minimum resistance (i.e. wiper nearest pin 7, IClb)

VR5 — to minimum resistance VR6 — to ground rail end VR7 — to minimum resistance.

Connect a signal to the input (probably best from a tape recorder phono output) and then connect the output to an external amplifier and apply power. First of all turn up VR1 and you should hear a totally unaffected signal. With VR1 set to maximum you should adjust VR5 so that the signal level is that you would have had without the unit in circuit. This sets the internal gain of the unit. Turn VR1 down and make sure that S1 is either open circuit or disconnected and then turn VR6 to maximum. If you are lucky you might hear the signal straight away but it is quite likely that you will have to adjust the pedestal biases for IC2. This is best done by adjusting VR2 a small way and then swinging VR3 from end to end until a sound is heard. You may have to go back and try a new setting of VR2 and adjust VR3 again. Keep doing this until you have a good clear signal and then make sure that both these presets are in the middle of the range that ensures a clear undistorted signal. Note: Adjust VR3 slowly as there is a time constant of a second or two from C17.

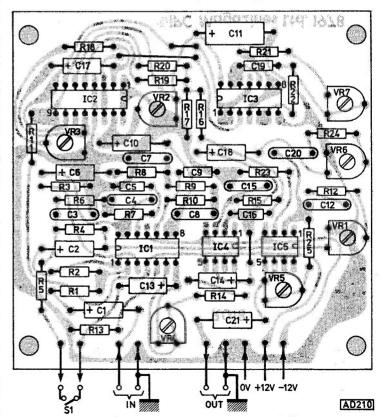
Once you have this signal you know that the bucket brigade circuit is working satisfactorily. While listening to the delayed signal try adjusting VR7 and you should hear a small change in frequency of music as you make the adjustment. This is a form of Doppler shift produced while you are altering the delay time. Wiggling VR7 (i.e. changing the clock frequency) will produce a vibrato effect. If, instead of using a simple oscillator, we had used a frequency modulated clock-say an NE555 suitably voltage controlled, we could have turned the unit into a true vibrato generator. Experimenters could try this by breaking the p.c.b. after pins 1, 2 and 3 of IC3 and applying a unity mark-space square wave to pins 3/14 and 8/10 of IC2. Note that complementary phases are required, i.e. a squarewave should be applied to pins 8 and 10 and its inverted form to pins 3 and 14.

You can now try to get the phasing effect. To do this adjust both VR1 and VR6 to mid positions so that you are mixing about equal proportions of raw and delayed signals. As you adjust the clock frequency you should hear the characteristic "whooshing" sound as you sweep through. The best effect is



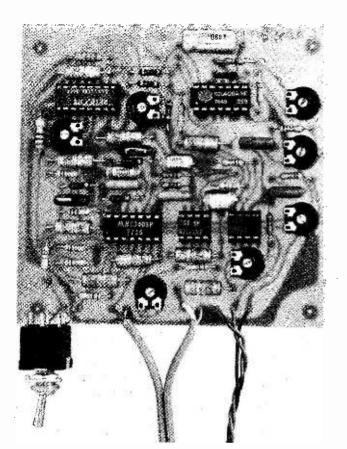
Fig. 2: Copper track layout (shown full size)

Fig. 3: The component layout



obtained when the two mixed signals are of the same amplitude so some experiment with VRI and VR6 will help.

Next try to get reverberation. First set VR6 to maximum and VR1 to minimum and close S1 (or short the respective board pins together). Set VR7 for maximum delay (i.e. maximum resistance) and



slowly increase the resistance value of VR4 (i.e. increase the feedback gain). For test purposes it is best to use a voice as the input signal. At some setting of VR4 you will hear a distinct "tinniness" occurring —at high settings of feedback it might sound as if someone were speaking down a pipe. This is a form of reverberation but probably not the sort you expected! Because of the short delay the feedback loop is more sensitive to some frequencies than others and this gives the rather metallic or coloured sound. Nevertheless careful adjustment of the controls will give some interesting effects.

For a better reverberation effect you should increase the delay and this can be done by increasing the value of C19. Try values up to about 560pF. Any values above this will reduce the clock frequency below that which will clash (alias) with incoming frequencies and an element of distortion will occur. Furthermore the post filter (IC4) will not be so effective in removing the switching glitches and you will notice an increase in background noise level (hiss). You can reduce this to some extent by increasing the value of C15 up to about 2200pF.

With long delays and S1 open you can get a form of "double tracking" or short echo which, when a person is singing, makes it sound as if there are two voices.

This unit is designed simply for experiment but we hope that many people will find a useful application. To get more realistic reverberation it is necessary to have a larger number of elements operating at different delays and feeding back into the same feedback loop—this gets rid of the colouration effect. In order to get a good echo effect (up to 500ms delay) over the whole audio spectrum it is necessary to have several more elements cascaded otherwise one has to use a clock that is so slow that aliased by-products are provided, giving rise to distortions.

INTRODUCTION TO LOGIC-6

Continued from page 46

in the counter chain into the display latches. This completes the measurement cycle. Usually a slower signal of perhaps 10 or 20 per second derived from the timing clock will be used to initiate the measurement sequence, by loading a new 1 state at the first stage of the shift register, so that a continuously updated display is produced. For making low frequency measurements, however, a manual push button may be used to start each measurement cycle thus giving a "one shot" mode of operation. Fig. 75 shows a basic system for timing and control logic.

The frequency of the clock applied to the shift register will determine the length of the count interval and hence the scaling of the display readout. With a 1Hz shift clock the readout will be in hertz, whilst with a 1kHz clock a display of kilohertz will be produced.

Input Circuits

The general arrangement of the input signal processing circuits is shown in Fig. 76. Here the signal is amplified and then shaped to produce pulses suitable for driving the logic. A prescaler counter may be included to simplify the design of the main counter by reducing the frequency of the signals applied to it. A prescaler will usually be required if v.h.f. or u.h.f. signals are to be measured, and here the prescaler is almost certainly going to be an ECL type device, such as the ones made by Fairchild and Plessey. These ECL counters are capable of handling signals up to 300MHz or even 600MHz.

The actual logic used in a digital frequency meter may vary considerably from what we have described here, but the procedures for choosing that logic scheme will follow the same general lines as those described above. In some cases it may help to draw up a timing diagram so that it becomes clear exactly how the logic sequence must operate.

Although the arrangement of logic gates for a given logic system can be worked out by applying common sense, it may result in an inefficient use of the gates available in the circuit packages. Remember there may be three or four similar gates in each package. By rearranging the logic it may be possible to reduce the number of packages required. For instance in a 7400 where only two of the NAND gates are being used, the other two could be used as simple inverters and might replace a partially used 7404 hex inverter.

NEXT MONTH—LSI AND THE MICROPROCESSOR



Burley p.s.u., November 1978

On the circuit diagram, C7 should read C4; on the p.c.b., C3 should read C5; chassis points (with the exception of that to the laminations of T1) should more properly be shown connected to the junction of BR1, C1 neg., C2 neg., and D1 anode.



R&C Bridge Kit

Kits can be categorised into two distinct groups. There are the complete variety which enables the constructor to build a finished piece of equipment without recourse to other sources of components or information and there are those which only provide the most difficult to make or obtain components.

The Philips Electronic Kits, released on to the UK market last year, are neither one type nor the other. On the one hand they provide all the parts necessary to make a working electronic device but stop short at providing any form of housing for the completed assembly. In fact no mention is even made of a suitable case for the unit.

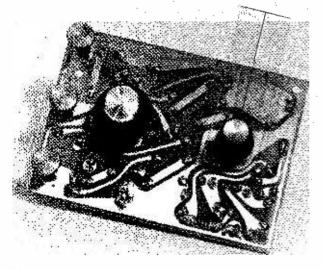
We decided to build one of the kits in the test instrument range and selected the Resistance and Capacitance Measuring Bridge.

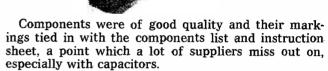
This instrument is claimed to be capable of measuring resistances from 10Ω to $10M\Omega$ and capacitors in the range $10 \mathrm{pF}$ to $10 \mathrm{\mu F}$. There is also an 'open-bridge' facility which allows the ranges to be extended or self inductances to be compared.

Components

The instrument is built on to one glass fibre printed circuit board and the components supplied fitted with no problems.

The instruction sheets provided with the kit are comprehensive and proved easy to follow. A separate sheet gives advice on how to solder for the complete novice.





The unit was simple to build and worked first time. However as a piece of test equipment it leaves a lot to be desired.

The layout of the controls is very cramped, mainly as a result of attempting to put all components and controls on the one printed circuit board. The three terminals supplied are not very satisfactory and the whole unit really needs rethinking as far as front panel layout is concerned.

A printed card panel is provided which fits the layout of the controls and terminals but makes it more difficult to fit the unit into a case.

Our kit had no knob supplied for the range switch, which is unfortunate as it should really match the knob supplied for the bridge control, which has a small shoulder moulded on to it accept the piece of thin Perspex sheet provided already ruled with the fine line pointer.

Accuracy

The instrument operates on conventional bridge principles and requires the operator to find a null, by listening through the earpiece for the point at which the audio tone is a minimum. This was found to be easy and is quite sharply defined. Using known value close tolerance resistors and capacitors to check the calibration of the scale proved that this was accurate. Two close tolerance resistors are provided with the kit to enable a simple calibration to be effected.

With a little more thought on the ergonomics of this instrument it would provide a useful addition to the home constructor's equipment. However it should not be difficult to rearrange the control and terminals on to the front panel of a suitable case.

This kit is one of a range recently introduced by the Philips Industries Group and the complete range is stocked by A. Marshall (London) Ltd., Kingsgate House, Kingsgate Place, London NW6.

Dick Ganderton





by Eric Dowdeswell G4AR

I have been surprised on several occasions when a reader has written in to say that he has been "off the air" because his set has been away for repair! It hadn't occurred to me that he would do anything else but find the fault and fix it himself! Apart from the cost of the repair the opportunity of gaining some useful experience in fault-finding is lost.

Naturally I would not recommend touching a set if it is still under guarantee but usually the problem is with a valved set which is much easier to fault-find on than a solid-state one.

The first essential is the receiver's manual which should be obtained when the set is bought, so that the user can study it and become familiar with the circuitry, and, incidentally, ensure that the set is being used to best advantage. Don't wait until a fault develops before obtaining a manual.

The second requirement is for a multimeter, preferably of $20k\Omega$ per volt sensitivity or better, so that it causes minimum upset to the operation of the circuit when it is connected to it. This is particularly important with solid-state devices. There are many good-quality imported multimeters on the market at reasonable prices but, like everything else, pay a little more than you can really afford and get some thing worthwhile.

Look for a large, open, uncluttered scale that is easy to read, and beware of the meter with a dozen scales reputed to be able to measure anything! If you want to measure anything other than voltage, current or resistance then use a separate instrument designed for the purpose.

If your pocket will run to a digital multimeter you will have the choice of a kit or a ready-built instrument. If you are not a skilled constructor then the latter should be your choice.

Most manuals have a fault-finding guide so study that carefully. Above all, be logical and don't go prodding around haphazardly or you will finish up with more faults than you started with. Remember that most receivers can be split into three sections, r.f. i.f. and audio, plus the supply lines. Analyse the

fault mentally and decide in which section it is likely to be and start from there.

Newcomers to the Column

Since starting this sub-heading a few issues ago I have been pleasantly surprised at the number of extra letters I have had from people who have been reading the column for a while but not bothered to write in. We all had to start somewhere so there is no shame in admitting to being a beginner, although not all who come under this sub-heading are beginners by any means.

Terry Wilson of Peterhead, Scotland, has been interested for a couple of years now and he and others are studying hard for the RAE under the aegis of local licensed amateurs. I was delighted to know that Terry's main interest is c.w. so perhaps we may get some c.w. logs into the column in future. From Beith in Ayrshire, Bill Steel writes to say his knowledge of amateur radio is zero! That's candid enough! He has a Sony RP8880 which allows him to copy stuff on 20m s.s.b., which is a good start. The Guide to Amateur Radio is also to hand now so you're on your way OM.

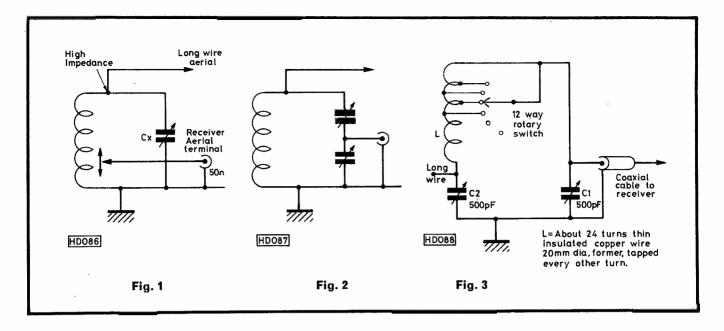
S. Norton of Garforth, Leeds, also says he is a complete beginner but with an FR50B and the PW Aerial Chart the first steps have been taken. I would mention that I send a long letter to all those needing advice on how to start in our hobby and I'm glad to say that many stay the course.

A bit further up the ladder we find C. A. Denton of Scarisbrick, Southport, who is an electronics technician but new to our game. He has a Marconi Electra and a Mercury plus a Redifon R146 and if anyone wants info on these rather rare sets C.A. will be very glad to help as he has the manuals. Write to him at "Hardacre", Pool Hey Lane, Scarisbrick, Southport.

From Seaford, Sussex, comes a note from Maurice Norman BRS37920 "a comparative newcomer" who has a DX160 receiver which he hopes to improve with the aid of the Aerial Chart, so here's looking forward to some reports soon OM.

Aerial Tuning Units

A reader recently reported that he was very disappointed with the results obtained when using a 66ft long wire with his new FRG-7 receiver "especially as it is a length that you seem to recommend". Well, I do usually add "plus an aerial tuning unit" (a.t.u.), and that is the crux of the matter. So let's try and elucidate.



The input impedance of virtually all communication receivers today is 50Ω , a low impedance compared to older sets like the HRO and AR88 which were around 600Ω , high enough to give results with a long wire connected directly to the aerial terminal.

Now the impedance of a long wire which is one or more half-waves long will be several thousand ohms. It is a matter of fundamentals that maximum power is transferred from one circuit to another when they have the same impedance, when they are said to be "matched". So if we stick the end of a long wire, which is high impedance, into the 50Ω aerial socket of a receiver we have about the worst mis-match imaginable! With a 66ft wire, which is a half wave at 40m (7MHz), the mis-match will be present on 40m and all higher frequency bands.

On 80m however, the wire is a quarter-wave long, presenting an excellent match to the receiver. An a.t.u. is not then essential, although I would recommend one in order to keep the system resonant.

In order to use an a.t.u. it is essential to understand how it works. The electrical equivalent of any aerial is a tuned circuit, and if one end of it is earthed (zero impedance) then the other end will exhibit high impedance at resonance, as does the end of the long wire, so the two can be connected together resulting in a good match.

Obviously there must be a point up the inductance where the impedance is 50Ω , to match the receiver input, if we can only find it! See Fig. 1. We can tap the coil every turn or so and use a switch for rapid selection of the right tap, indicated by a significant increase in signal strength, simultaneously keeping the system resonant using tuning capacitor C_x .

To avoid tapping the coil we can split the capacitor into two parts, Fig. 2, to form a capacitative divider but retaining sufficient effective capacitance to maintain resonance. When the capacitors have equal values we are virtually centre-tapping the coil.

For practical reasons the circuit is rearranged, Fig. 3, into the familiar arrangement where C1 is used for "loading" and C2 is used to maintain resonance. The coil is tapped to provide optimum inductance on each band. All the turns will be required for the 160m band and two or three turns for the 10 and 15m bands.

Initially, set C1 about half way and adjust the number of turns until C2 will tune the particular band. All three adjustments are inter-dependent for best results so a fair amount of juggling is called for. Once the optimum settings for each band have been found write them on a label and stick it on the a.t.u. It should be possible to resonate almost any length of wire on any h.f. or l.f. band if the a.t.u. is working properly.

Round the Bands

Bob Bell (Blyth, Northumberland) found 10m quite active after a period of quiet, with west coast Americans aplenty, while Bill Rendell carried on his search for new island prefixes from Truro in Cornwall. His old valved AR3 plus preselector is fed from folded dipoles on 10 and 15m but 20m is not neglected. Bill found JW7FD, as did others, from Bear Island for a new one. Seems JW has a population of just 13! J3AH on Grenada turned up again on 15m plus YB0CR on Java, with KB6EI on Baker Is on 10m to keep the pot stirring.

In Crowthorne, Berks, **Allan Stevens** has been working his *PW* Direct Conversion set on 20m and logged some 54 countries in the first couple of weeks, but he has got a PR40 preselector in front, with a half-wave aerial indoors. Best catch was HH2SD on Haiti. **Ian Marquis** A9140 doesn't get much time for listening with all his school work to be done but promises to have a go on the l.f. bands this coming

Reports on the various bands are welcome and should be sent direct, by the 15th of the month, to:-

AMATEUR BANDS Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey KT21 2TW. Logs by bands, each in alphabetical order.

MEDIUM and SW BANDS Charles Molloy G8BUS, 132 Segars Lane, Southport, PR8 3JG. Reports for both bands must be kept separate.

VHF BANDS Ron Ham BRS15744, Faraday, Greyfriars, Storrington, Sussex RH20 4HE. winter. CE0AE in Antarctica was a good one as was KG6JJH on Guam on 10m s.s.b. Ian has an FRG-7 with a 70ft wire plus a 16ft vertical and a 10m dipole.

Dave Garner BRS39645 writes from Newhaven, Sussex, to say that although he has been interested in s.w. for a number of years he is only now starting to take it seriously, studying for the RAE next year. In an all-night stint he managed 49 countries on his FRG-7 on 14MHz with a loft wire. A converter for 2m is fed by a PW Slim Jim aerial. Dave mentions 4X30JU a special call celebrating 30 years of Israeli independence. Dave very kindly donated his old R107 receiver to one of our young listeners and it should have reached its new home by now. Any other offers are welcome.

General Notes

Although Dave Parker of Elstead, Surrey, has an HQ120X receiver he still plays around making sets and has at last got one to work, but it lacks selectivity. I always advise people to get a set of some kind first of all and to play around afterwards, otherwise failure to get a set working can be very disheartening. It's nice to have a working set as well, just to prove that the bands aren't all dead! Dave has joined the RSGB but comes up with the eternal complaint about Radio Communication "there is nothing in it for the beginner".

Joe Porter in Belfast has 140ft of wire up in the air but reckons it is no better than a short vertical. Since it is going up and down the garden a few times I suspect the signals on the long wire are cancelling themselves out! An aerial tuning unit, as described

below, might be a good idea, too Joe.

From Reading, Berks, Paul Bown BRS40740 reports a lack of success with his Skywood CX203 mainly because some of the r.f. wiring is missing! He's about to scrap it for spares and buy a new set. Might be

the best in the long run, OM.

An appeal from Vic Tuff of 38 Fourth Avenue, Blyth, Northumberland, who is electronics instructor at the local ATC squadron. Vic takes old equipment and refurbishes it for use by the squadron and wonders if any reader has old service equipment or handbooks or manuals. Whatever the gear, it will be either used or put into a museum, after overhaul. Vic is also associated with the Blyth radio club that has recently been started with reader Bob Bell as one of the leading lights.

Martin Leizers of Newport writes to say his new call is GW8RKB and he is on the air with a borrowed Yaesu FT220 and the Slim Jim aerial although he has not neglected to listen on the h.f. bands, with his Realistic 160 and 150ft of wire, as his log extract

shows.

Club News

The Thames Valley ARTS welcomes newcomers to its meetings on the first Tuesday of each month at Giggs Hill Green Library, Thames Ditton, Surrey. December 5th "promises a wealth of stories and experiences" from Dave Foster G3KQR, a member of the police force it seems! More info from Sec R. Blasdell G3ZNW, 92 Bridge Road, Chessington, Surrey.

The Cheltenham ARA should have its AGM on December 7th but I understand it will now be in January so check with Sec G3JJG QTHR.

Log Extracts (All s.s.b.)

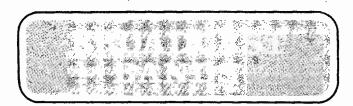
M. Leizers:—40m HI8RRD HK5BKI HR1JMV; 20m EA9FD KC4AAC KZ5ED VP2LFD VP8QI; 15m HC5EA HK9KL: 10m HK3AXT KH6IBA.

J. Porter:—15m VP2MAY; 10m 8P6JQ.

- D. Parker:—40m ZL1DNQ ZL4BE; 20m CO2FA; 15m CE9AT.
- D. Garner:—20m CT3AB JW7FD VE6BGU/SU 4X30JU.
- I. Marquis:—80m EA8OZ EL2T 7X5AB; 20m CE0AE; 15m KH6HGP; 10m CT2BB CT3BX KG6JJH VS6FI.

A. Stevens:—20m HH2SD ZD7SS.

W. Rendell:—20m JW7FD OY5G SU1CR VE8RR VK9XW VP2LLF XT2AE; 15m J3AH JW7FD TU2HS YB0CR; 10m EA8RG KB6EI VU2GAA.



MEDIUM WAVE DX by Charles Molloy G8BUS

This issue of PW marks the tenth anniversary of my first contribution to the Medium Wave Column. At first I wondered if anyone read it as it was some time before the first reader's letter arrived. Since then several hundred letters from all over the world have been received, evidence that interest in the medium waves is far from dead. The hobby of DXing began on the medium and long waves with the start of broadcasting itself in the early twenties. Magazines of the day not only reported North American DX but also carried reports from readers who actually listened to the programmes. Amateur Wireless of November 1st, 1924, had a full page article called "A great night with WGY", which was on 380m. WGY Schenectady, New York, is still on the air on 810kHz, but is a lot more difficult to pick up than it used to be. Receivers are more powerful and selective but interference is considerably more severe and this is the challenge the band offers to the DXer-how to winkle out the DX. There are more broadcasting stations on the medium waves than on all of the short wave bands put together, a fact that most people including myself, find surprising.

Long Wave DX

Although the frequencies allocated for use on the long waves are unaltered under the new Geneva Plan, there will be quite extensive changes in channel occupancy. Incidentally, 151kHz, 180kHz, 186kHz and 251kHz were not authorised under the old plan (nor are they in the new one) and the stations operating on them should really be classified as pirates!

A number of new countries will appear on the band on or after November 23rd. Egypt is authorised to use 164kHz and 200kHz, Spain will have Madrid on 191kHz and Barcelona on 227kHz, Libya will have 236kHz, Italy will have 245kHz, Eire (Tullamore) and

Syria will share 254kHz, Bulgaria will have Plovdiv on 263kHz while Tel Aviv in Israel will be authorised to use 281kHz. After a long absence Holland will come back on the long waves with Lopik on 173; at one time Huizen was on 223kHz.

What will happen to Luxembourg? This country was not authorised to be on the long waves under the old plan and it isn't under the new one either. Radio Luxembourg started its English transmissions many years ago on 1293 metres, providing an interesting alternative on a Sunday evening to chamber music and the epilogue. The BBC are authorised to use both 200kHz and 227kHz and it was originally intended to use both channels for BBC Radio 4 with 200 for Droitwich and 227 for a new site in Scotland. The latest list from the BBC makes no reference to 227kHz so it looks as if plans for a second long-wave outlet have been shelved or abandoned.

Altogether, the long-wave band should become a more rewarding place for the DXer under the Geneva Plan. The 15 channels will be occupied by a large number of interesting broadcasters and once the full potentialities of the band are realised then it will only be a question of time before someone works out a design for a standard long wave loop.

Medium Waves and the lonosphere

Does the weather affect medium wave DXing, asks reader Malcolm Lougharne and the answer must be that this is unlikely. The weather is confined to a region close to the earth's surface while a DX signal travels much higher, up to the ionosphere, spending the greater part of the journey well above the weather. Paradoxically, weather may have an effect on the ground wave. One can visualise a long wet spell (English summer) or a drought, producing a change in ground conductivity.

The existence of an ionosphere was suspected when broadcasting first began though there was an alternative view that a long distance radio signal just followed the earth's surface. It was left to Appleton and Barnett to prove that a reflecting layer did exist. This happened late on December 6th, 1924, when they conducted an experiment using the BBC transmitter at Bournemouth (780kHz) and a receiver at Oxford which was equipped with a signal strength meter ("S" meter).

The signal from transmitter to receiver travelled by two routes. The shorter route was the ground wave and the longer the reflected wave. The total signal applied to the receiver would depend on the phase relationship between the two. If the two signals were in step (crest of one corresponding to the crest of the other), then the total signal would be a maximum. If they were out of step (out of phase) then the total signal would be a minimum. By slowly adjusting the frequency and hence the wavelength at the transmitter, the two signals at the receiver would move in and out of phase with each other producing alternating maxima and minima on the "S" meter, providing of course that there really was a reflected wave. This is what happened. Moreover, by measuring the wavelength at two adjacent maxima it was possible to calculate the height of the reflecting layer which turned out to be about 60 miles above the earth's surface. It was called the Heaviside Layer after the man who predicted its existence but later on it was renamed the E Layer.

It is hard to believe that this elegant experiment took place within living memory. It is even more surprising to read in some current literature that the E layer disappears after sunset! This clearly is not so. The critical frequency of the E layer falls to about 500kHz some four hours after sunset and remains at this value throughout the night. This means that the highest frequency returned from low angle radiation (a DX signal) will be in excess of 2MHz and consequently it is the E Layer that is responsible for most of the DX to be heard on the medium waves. High angle radiation can penetrate the E Layer but that is another matter altogether.

Readers' Letters

Harold Emblem (Mirfield) reports that Asiatic DX has been good recently. Stations logged with his Eddystone 730 and loop were Quetta Pakistan on 750kHz (756), Hyderabad on 1010kHz (1007), Rajkot India on 1070 (1071). The frequencies in brackets are those allocated for use after November 23rd.

How should I connect a loop to my DX160 asks Mark Hallam. This receiver has terminals marked A1, A2 and GND. The loop should go to A1 and A2 and any link between A2 and GND should be removed. GND should go to earth. When using a long wire connect it to A1 and join A2 to GND which should be earthed. It is not necessary to use coaxial cable between the loop and the receiver. Plastic lighting flex of the type where the wires run parallel to one another will do instead, as it is a reasonable substitute for 750hm flat twin feeder.

Harold Brodribb (St Leonards-on-Sea) complains about the BBC European Service transmitter at Crowborough (809kHz) which has been putting out a strong harmonic on 1618kHz. On the long waves, Radio Finland now carries a news bulletin in English over its transmitter at Lahti on 254kHz at 2015. Finally from the Isle of Wight comes a note from William C. Savage, who says that he received QSLs from WHAM Rochester NY, WBZ Boston and KDKA Pittsburg away back in 1928 and he wonders if anyone can predate this.



SHORT WAVE BROADCASTS by Charles Molloy G8BUS

A letter from reader Roy Haynes in the October issue mentioned five countries that cluded him no matter how hard or how often he tried for them. Four of these are now reported by readers. Mark Hallam (Hereford) used a Realistic DX160 and an indoor aerial to pull in Radio Afghanistan in English at 1900 on 11820kHz, with slight QRM from Radio Moscow. The SLBC (Sri Lanka Broadcasting Corporation) was also heard by Mark on two frequencies with programmes in English. The All Asia Service was logged at 0100 on 15425kHz and again at 1400 on 9720. Reception on both channels was poor; not good

enough for a reception report. Jeddah in Saudi Arabia is reported on 11855 at 1910 in English by Noel Cosgrave (Dublin) who was using a Mullard MAS 1659 receiver and a 75ft long wire. Jim Edwards (Bryn, near Wigan) has an FRG-7 and a 50ft long wire and he picked up Dacca in Bangladesh on 15285kHz at 1815. Only one of Roy's difficult countries remains. Has anyone heard Syria?

Keeping a Logbook

Why go to the trouble of keeping a logbook? Well, as a record of stations heard it should add to the interest and pleasure of DXing and for that reason alone it is worth while keeping one. A station log will also be of practical value for it will be a record, for future reference, of information about reception such as the best time of day or year for listening to any particular area. If the receiver has a logging scale then the reading for each station can be recorded in the logbook making it easy to tune back to a station or programme. A written record is more reliable than memory when it comes to writing a reception report.

What form should the logbook take? The DXer can please himself. I use a 9in by 7in stiff-backed notebook which has horizontal lines printed on each page. Each line is used for one entry in the log and I pencil in vertical lines to divide entries into the number of sections required. These are, starting from the left: date, time in GMT, frequency in kHz, log scale reading, programme heard, SIO rating, station heard (if identified), aerial used. The DXer can add to or subtract from this list to meet his own requirements. The date is the date at the DXers QTH; it may be different at the transmitting end. The time should be GMT as the use of summer time can cause confusion. I prefer SIO as an indication of the quality of reception but other codes such as SINPO, SINFO or SIFO can be used instead and even Excellent, VG, Good, Poor will be of value. Fill in the logbook while you are actually DXing and it will soon become a habit.

Readers' Letters

I would like to contact someone of my own age group, female, and who is interested in s.w. listening and DXing, writes Noel T. Cosgrave (wouldn't we all!). Noel, who is 14, lives at 102 St Josephs Road, Greenhills, Dublin DN12, Eire. There were two lady DXers among the first forty in the Sweden Calling DXers Jubilee Contest and a Norwegian DXer, Grete Osmundsen spoke recently on DX Juke Box (Radio Nederland) about Women in DXing. This brought replies from more than 40 women DXers who were in the age range 14 to 66, so you might be lucky Noel.

A problem concerning the Murphy B40 ex-WD communications receiver comes from Roderick Williams who has been unable to get a replacement output valve (EL22). He has managed to get the set working using the pentode section of an ECL86 but he wonders if any reader can suggest a better alternative. Has anyone any ideas?

Aerial Wire

Fairly heavy gauge copper wire is usually specified for aerials and while this type of wire will certainly give good results, especially when used for transmitting where currents of significant value are involved, it has some disadvantages for broadcast band DXing. Copper corrodes, so bare wire is not really desirable. My aerials are subjected to the full force of gales from the Irish Sea and I prefer something a bit stronger than copper. For several years I have been using the plastic covered steel wire that is on sale in gardening shops, which can be purchased in drums as well as in hanks, well aware that I might be sacrificing something to obtain mechanical strength. Recently a "small ad" in PW which seemed to indicate a solution caught my eye. "Aerial wire, 20swg, copper plated, steel core, tough pvc insulation". My 90ft long wire was due for overhaul anyway so it now has new wire and insulators and it has already withstood some rough autumn weather.

Why copper-plated steel wire? Skin Effect is the answer. Direct current spreads itself evenly over the cross section of the wire. So does low frequency alternating current, but as radio frequencies are reached there is a tendency for the flow to migrate to the surface of the wire as a result of interaction between the current and its own magnetic field. At very high frequencies nothing at all passes along the centre of the conductor and a tube can be used instead of wire. Steel is not such a good conductor as copper but it is mechanically strong so this is the material for the core. Copper is a good conductor so this is the stuff for the surface of the wire where the current density is high. The plastic covering protects the copper from the effects of the atmosphere and it also acts as an insulator if you want to use this wire for a downlead.

I have not noticed any spectacular improvement in reception since using the new aerial and I did not expect it, but I would hope for a weak signal to show an improvement and perhaps become resolvable. The new wire should be more effective on the higher frequencies which will be very welcome now that they are opening up. AMTEST, 55 Vauxhall Street, Worcester WR3 8PA, supply this type of aerial wire and they were advertising in PW at the time of writing.

DX and **News**

More news of DX on the 11m band comes from Harold Brodribb, of St Leonards-on-Sea, who reports hearing Radio Tirana Albania on 25290kHz (out-of-band) on several occasions between 1700 and 1800 using his AR88LF and a loft multiple dipole. George E. Lee (Osset, West Yorks) reports that Radio Australia on 11800kHz at 1800 is a very good signal, better than on 9570 at 0700 which is intended for reception in the UK. David Wyatt (Oswestry) has built an a.t.u. for use with his home-brew 6-transistor receiver and this makes a tremendous difference when connected to his 100ft long wire. Recent DX includes the Voice of Chile on 17800 at 2200.

A Vega Spidola connected to the TV aerial is the rig employed by Andrew Rogers (Bristol), who reports hearing Radio Grenada on 15105 at 2000 while Mark Hallam has picked up Radio Yugoslavia on 9620kHz at 2200 which has eluded him for some time. A number of out-of-band transmissions are reported by Bob Bell of Blyth. These are India on 11620kHz at 2015, Pyongyang on 11535 at 2010, Peking on 11445 at 2005, Pakistan on 11675 at 1745 (not listed) and USSR on 25590 (probably a harmonic). There are quite a few stations to be found in between the standard s.w. bands and it can be

an interesting diversion to dig out some of them. Broadcasting out-of-band is likely to increase, partly as h.f. bands are being used less for commercial use and partly in anticipation that the World Administrative Conference 1979 will award additional space for broadcasting.



by Ron Ham BRS15744

There should be a law against DX coming at meal times, or during working hours, as did the extensive aurora which manifested during a Friday afternoon in September, or the super-strong signals from Japan, on 10m, during breakfast and the crop of international beacon signals at lunch time. However, through the vigilance of my readers, each making an observation when possible, I can piece together an interesting and scientifically valuable report.

Aurora

The intense aurora borealis which began around noon on September 29th was brought to my attention by Alan Baker, G4GNX, Newhaven, Sussex, who was on holiday at the time. During the event, which ended about 1730, John Branegan, GM8OXQ, Saline, Fife, used a hand elevated 8-element Yagi to establish the elevation angle for maximum auroral signals. Although his results suggested that the auroral reflector was about 15 degrees from his QTH, John finds this hard to explain because of the strong signals on 2m coming from France and Luxembourg. At the other end of the UK, Alan Baker found that his beam headings were 020 to 070 degrees with a peak at 040.

I cannot add to this because I monitored the early part of the event with a vertical dipole feeding my R216 receiver, from which I heard signals from several mobile radio stations around 72MHz, about 17 continental carriers in Band II and some 30 rough signals between 50 and 75MHz, all bouncing off the aurora. John Branegan heard his own 29MHz OSCAR-8a down-link signals go auroral and heard tone-A signals from the 2m beacons in Cornwall, GB3CTC, Northern Ireland, GB3GI, Lerwick, GB3LER and Wrotham, GB3VHF as well as from amateur stations in DJ, EI, F, G, GI, GW, LX, ON, PAO, and SM. Down south, Alan Baker heard signals from GI, GM, GW, LX, SM and SP, worked his first EI on 2m, EI5BH, G2DUP, Cornwall, GW2HIY, Anglesey, and G4CJG, Durham, who was the most consistent auroral c.w. signal in the south. Alan also worked G4CLA, Manchester, whose auroral s.s.b. was the strongest in Sussex, but his real prize was an auroral QSO with his friend, Roy Bannister, G4GPX, in nearby Lancing, Like many others Roy had a good haul of DX during the event. Less prominent auroral disturbances took place around 1700 on September 25th and 27th.

The 10 Metre Band

One of the most consistent signals on 10m between September 20th and October 18th was A9XC, the International Beacon Project station at Bahrain, closely followed by the Cyprus beacon, 5B4CY. John Branegan also heard A9XC consistently and 3B8MS in Mauritius occasionally. Like myself, Alan Baker, Gordon Goodyer, BRS 37345, Petworth, Sussex, Harold Goble, G4FDQ, Lancing, and Graham Lay, West Chiltington, Sussex (a new reader) using an AR88D receiver, have all noted the good conditions on 10m during the first 3 weeks in October, bringing extra strong signals from the far east during the early mornings and the same from the Americas, Canada and many other countries during the afternoon and early evenings. Towards the end of September Harold Brodribb, St Leonards-on-Sea, heard North American CB stations calling their counterparts in W. Germany and at one time thought W5UAW, Mississippi, was going to burst his loudspeaker. Martin Liezers, Newport, Gwent, reminds me that in mid-August 10m was also open to S. America until about 2000 on several days, so the band has perked up consistently in recent months.

Solar

John Branegan is now making regular observations of the sun and like Cmdr Henry Hatfield, Sevenoaks. Kent, John Smith, Rudgwick, Sussex, and myself recorded solar noise between 136 and 147MHz on September 22nd and 23rd. From 1130 to 1330 on the 23rd, D. S. Jones, GW3XYW, Pontardulais, Swansea, using an 18ft dish aerial noted a large increase in solar noise at 432MHz while he was testing his EME gear which I hope to hear more of in the future. Henry Hatfield, using his spectrohelioscope counted 24 sunspots on September 21st, saw the remains of a ribbon flare on the 23rd, a bright plage on the 29th, about 20 spots in 8 groups with lots of tiny filaments on October 12th and the remains of a ribbon flare on the 13th, following a large burst of radio noise at 1237 which lasted for about an hour. Solar noise was also recorded on October 3rd, 5th, 9th, and 11th.

Sporadic-E

In our November issue I asked for gen about a caption which read "Granada Television International" well, Frank Luman, Glasgow, says "It is most likely RTVE, from Madrid (E2). Many BBC and ITV programmes go on to Spanish TV. I've seen the "Muppets", 'Starsky and Hutch' and a situation comedy put out by Thames TV". Frank agrees that things are quieter now the sporadic-E season has finished and reminds me that bands II, III, IV and V are affected by tropospheric disturbances and require high gain systems and pinpoint rotation of the aerial to resolve the DX. Like Frank and his fellow DXers in Scotland, I am looking forward to some good tropo openings this winter.

OSCAR

Our satellite specialist, John Branegan, says "The QSL cards for my first two months of satellite work are now rolling in, 41 in one week with 10 countries confirmed. Best DX K9KB, Fort Wayne, Indiana, but

best of all, DD1QS, confirming my first QSO via OSCAR's mode J". To date John has worked 23 countries on mode J, the most recent being CT, LX, and OH.

Microwaves

As usual, Ern Downer, G8GKV, Worthing, and Ern Hoare, G8BDJ, Southwick, Sussex, were operating portable on Chanctonbury Ring, near Worthing on October 1st for the fifth and final leg of the RSGB's 10GHz Cumulative Contests for 1978. They spent most of the day in low cloud and the water was streaming off their dishes. At 0900 the Alderney 10GHz beacon was about 30dB above the noise but gradually dropped away during the morning and disappeared completely by early afternoon.

They both worked two new stations, G4ETU/P and G8JVE/P also on the South Downs, two home based stations, G3JVL and G8DIC in Hayling Island, G3IFF/P near Portsmouth and G3JHM/P near Alton, Hants. The two Erns also heard signals from F6DLA/P but conditions were rock-bottom for 10GHz and no complete QSOs were made with French stations. Ern Downer said "although we have survived some atrocious weather during the five legs it will not deter us from going portable whenever anyone wants to carry out tests".

Tropospheric

During the late evening of September 21st I heard continental signals coming up in Band II and by 0700 on the 22nd it was obvious that a tropospheric opening had begun. While I heard strong signals from GW mobiles through the Bristol Channel repeater, R6, French stations through the Kent repeater, R4, strong pictures from Lichfield on Channel 8 and a 539 signal from the Sutton Coldfield beacon, GB3SUT on 70cm, Martin North, G8HKK, Bath, had a 59 f.m. QSO with G4GNX, Newhaven, on S16. At 0700 on the 23rd, French stations were working through the Hampshire, GB3SN, Kent, GB3KR and Malvern Hills, GB3MH repeaters, strong pictures were coming from Lichfield, many French broadcast stations were audible in Band II and while Band V TV was upset, GB3SUT was 559 on 70cm in Sussex. During the late evening, John Cooper, G8NGO, Cowfold, Sussex, and G4GNX worked into France on 2m and G4GNX also worked GJ and, to his delight, LX. Early on the 24th, G4GNX, G8NGO and David Coble, G8HDF, Lancing, worked into DJ, and David, using a Liner-11 with an HB9CV mini-beam in the loft also had QSOs with stations in GU, ON and PE. Changes in the atmospheric pressure coupled with the remarkable fine weather unsettled the v.h.f.s from October 10th to 14th, because, early on the 10th, Mike Gaskin, Croydon, heard Hilversum 1 and 2 and several German broadcast signals in Band II and on the 12th Continental stations were again heard between 88-100MHz by Ian Rennison, Horsham and myself. Peter Penfold, West Chiltington, Gordon Goodyer and Alan Baker all commented about the large number of French stations working through the southern England repeaters. Guy Stanbury, Chelmsford, Essex, is now in contact with John Ding, Suffolk, Rodney Sixe, Cornwall, Des Walsh, Co Tipperary, and Martin Warburton, Wiltshire, who, like myself, are all Band II enthusiasts and I shall look forward to hearing from them all in the future.

Club News

On September 21st, the Sussex Repeater Group held a mammoth junk sale in Brighton in aid of funds and the organisers were overwhelmed with the results and wish to thank everyone who supported them.

Fifteen members of the Brighton and District Radio Society visited Cmdr Henry Hatfield's observatory on September 30th and were fascinated by the intricate workings of Henry's spectrohelioscope.

Several of our readers attended the first open day of The Chalk Pits museum at Houghton, Sussex, on October 15th and looked in at the site radio building where the early radio collections of David Rudram, Worthing, and myself are on display. This new industrial museum, which should open next May, covers more than 30 acres, so, to assist the organisers and the thousand or so visitors, Cadet Wireless operators with six v.h.f. sets from Flt/Lt John Keegan's 2464 (Storrington) Squadron, ATC, maintained the day's communications.

The Chichester and District Amateur Radio Club have conferred the title of Honorary Life Member upon one of our original readers, Russell Ward, G8AFT, for past services to the club.

Russell, now 78 years old, became interested in radio in 1922 when he was a member of the first ever radio club at RAF Cranwell. Although he left the RAF in 1924, he remained on the reserve and was recalled during World War 2, serving as a Flt/Sgt Electrical Fitter.

Unfortunately, through ill health, G8AFT spends much of his time in bed, so fellow club members have installed a 2m transceiver by his bedside to keep him in touch with his many friends in the world of amateur radio.

From Down Under

Anthony Mann, Applecross, Western Australia, says "August was rather poor, with nothing above 41MHz. A rapid improvement was noted since early September with 36MHz US v.h.f. in almost daily, and Far East m.u.f.s in the 45-50MHz region. Consequently E2 Malaysia and R1, 49.75MHz, China/USSR have been seen several times". Anthony received strong pictures on Chinese R1 between 1700 and 1800 on September 16th, 1500 on the 17th, and 1700 to 1900 on the 18th, but on the 30th, Chinese R1 was seen around 0930 which was the first time that Far East DX has been in so early. On September 29th, signals from the Hawaii 6m beacon, 50.104MHz, were received in Adelaide, S. Australia, and during the morning of October 1st, there was a strong 52MHz opening to Japan.

Thank you all for your most interesting letters and the support you have given my column throughout the past year, I wish you all a merry Christmas and look forward to hearing from you again in 1979.

Stop Press

Some super long-distance v.h.f. reception is reported in a letter just in from Anthony Mann in Australia. Channel B1 BBC TV sound on 41·5MHz was received there on October 13, 14, 16 and 19th with the companion vision channel on 45MHz also present on the 13th. French Channel F2 sound on 41·25MHz was heard on October 16th.

FIt. Lt. JOHN KEEGAN

by RON HAM

At the age of 14, John Keegan's interest in radio was stimulated by the variety of signals he heard on the short-wave bands of a domestic broadcast receiver. Very soon he became a keen home constructor and well remembers the thrill of his first home-brew receiver, comprising a 1T4 valve, Repanco coils and a pair of headphones which he later replaced with an amplifying stage and a loudspeaker.

In 1962, John, then living in Liverpool, became a Civilian Instructor with 611 Squadron, Air Training Corps, and used the famous R1155/T1154 equipment on the ATC's national h.f. network. The ATC became a major part of his life. In 1966 he was commissioned as Pilot Officer and in 1968 was promoted to Flying Officer. Later, moving to Nottingham, he joined 1359 Squadron and served there for two years.

John is a regular reader of Practical Wireless and has a complete set dating back to 1961. His interest in v.h.f. radio, fostered by the ATC, is maintained in the air, because, as a civilian pilot, he holds a v.h.f. aeronautical operator's licence. In 1974 he moved to Steyning, to take up the post of Principal Develop-



ment Engineer, Human Factors, with a Sussex firm who build flight simulators, and the ATC posted him to 2464 Squadron in nearby Storrington. Promotion to Flt. Lt. and to Commanding Officer of 2464 Squadron followed in 1976, since when he has devoted a major part of the Squadron's training programme to v.h.f. radio.

In the picture with John Keegan (back left) is Flt. Sgt. Maynard, one of the Squadron's senior radio operators, Cpl. Nigel Golds (back right) a keen broadcast and short-wave listener and (seated) Cadet David Sopp, a trainee wireless operator.

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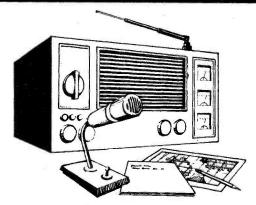
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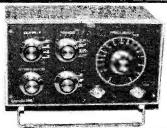
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Mini BIMDRILL with 3 collets up to 2.4mm dia £ 8.10 Major BIMDRILL with 4 collets up to 3mm dia £13.60

Accessory Kits 1 have appropriate drills and collets as above plus 20 assorted tools. Mini Kit 1 — £15.12, Major Kit 1 — £19.44. Accessory Kits 2 have appropriate drills, collets plus 40 tools and mains 12V dc adaptor. Mini Kit 2 - £34.02, Major Kit 2 - £39.42.

Accessory Kits 3 as appropriate Kits 2 plus stand/lathe unit. Mini Kit 3 - £45.36, Major Kit 3 - £50.76.

BIMDIPS



Rapidly inserts and withdraws any 4-18 pin, .3" pitch DIL package without beding the legs. Adjustable metal jaws for MOS type devices grip the bottom of the leg for minimum strain. Will

pick up IC's from a bench, a carrier or a pcb. £13.77.

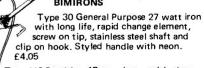
BIMSNIPS



Precision made side cutters, spring action, ground steel fine pointed blades for intricate work

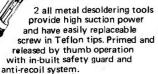
514" tong £3.34

BIMIRONS

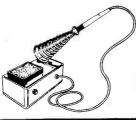


Type M3 Precision 17 watt iron, quick change tip, long life element, styled handle with clip £4.43

BIMPUMPS



BIMPUMP Major (180mm long) £7.99 BIMPUMP Minor (150mm long) £6.80



BIMSTATION

ype PSU6 Soldering Iron Station complete with 6V, 6 Watt miniature iron having stainless steel shaft, quick change slide on tip and long life element.

Station contains 240V/6V transformer, neon, coiled iron support and sponge iron tip cleaning pad.

New product available shortly

on hook

BIMDICATORS



.125" dia, lens



ECONOMY QUALITY LED's

Mixed bags of .125" and .2" dia. lens in various colours 50 for £5.67, 100 for £10.00

FULL SPECIFICATION LED's

.125" or .2" with mounting clips and data Red -£1.67/pack of 5, Green -£2.48/pack of 5, Yellow/Amber -£3.18/pack of 5



33 and 34 SERIES

Front viewing (30° angle) LED indicators

BIM 33 is nickel plated, uses 3.2mm dia LED and needs 6.5mm dia. fixing hole.

BIM 34 is chromium plated, uses 5mm dia. LED and needs 8mm dia, fixing hole.

Red - £2.80/pack of 5, Green/Yellow - £3.24/pack of 5





AH A SERIES

240V Neon with integral resistor. held in 8mm hole by plastic bezel.

Red, Amber, Clear or Opal lens £2/pack of 5, Green lens £3/pack of 5 Low Voltage equivalent of above with Red, Amber, Clear, Opal or Green Lens. 6V £0.54 each, 14V £0.58 each, 28V £0.65 each

State Voltage, lens style, colour and whether tags or flying leads.



D SERIES

LES and Midget Flanged lampholder with 13mm dia. (A) and 18mm dia (B) lens. Solder tags. ½" dia. hole fixing (lamps not supplied) plus chrome bezel with A lens.

Red, Amber, Clear, Green, Opal £0.66 each



G SERIES

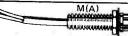
TI Midget Flanged lampholder. Lamps are available on request. 8mm fixing hole, solder tags. Front replaceable, 7.25mm dia.

lens. Red, Amber, Clear, Green, Opal £0.43



05 SERIES

240V Neon with integral resistor. Self retaining in 13mm hole, Solder/,25" push on blades. 13mm dia. lens with 19mm dia, chrome bezel. Red and Amber £0.61 each, Green £0.78 each.



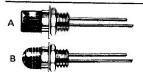




M & MP SERIES

Low voltage nickel plated brass (M) and Polycarbonate (MP) indicators, 150mm leads, 6.4mm fixing hole Red, Amber, Clear, Green, Opal

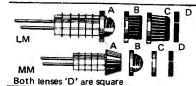
6.9mm dia. lens (M) 6V £0.65 each, 14V £0.68 each, 28V £0.79 each 7.5mm dia. lens (MP) 6V £0.55 each, 14V £0.59 each, 28V £0.68 each 6.9mm dia. lens (M)



BIM M LED SERIES

Nickel plated brass bodied LES indicator, 21mm wire wrappable leads, 6.5mm fixing hole, 2 styles, 6,8mm dia

Red £0.67 each, Green £0.83 each, Amber £1.00 each



BIM LM & MM LED SERIES

Subminiature nylon bodied LED indicators with 12mm wire wrappable leads

LM & MM push fit into 4.75mm & 4mm holes respectively. Each series has 4 lens styles in Red £0.67, Green £0.83, Yellow £1,00 each.

23 26 56

BIM 23, 26 & 56 LED SERIES

Black nylon bodied LED indicators. BIM 23 has 7mm flat face, BIM 26 & 56 utilise 4 & 5mm dia LED's, Push

fit in 8mm hole. Red £0.46 each, Green £0.62 each, Yellow £0.77 each

BIMACCESSORIES



BIMDAPTORS

Allows pcb's to be flat mounted sandwich fashion in BIMBOXES BIMCONSOLES, and all other enclosures having

1,5mm wide vertical guide slots. One plastic BIMDAPTOR on each corner of pcb(s) enables assembly to be simply slid into place. 54mm long, 10 slots on 5mm spacing and can be simply snipped off to length,

Packs of 25 £1.08 per pack

BIMFEET



11mm dia, 3mm high, grey rubber self adhesive enclosure feet.

Packs of 24 £0.77 per pack

BIMBOARDS



COMPATIBLE BIMBOARDS



ccept all sizes (4-50 pin) of DIL IC packages as well as resistors, diodes capacitors and LEDs. Integral Bus Strips up each side for power lines and Component Support Bracket for holding lamps, switches and fuses etc. Available as single or multiple

units, the latter mounted on 1.5mm thick black aluminium back plate which stand on non slip rubber feet and have 4 screw terminals for incoming power.

BIMBOARD 1 has 550 sockets, multiple units utilising 2, 3 and 4 BIMBOARDS incorporate 1100, 1650 and 2200 sockets, all on 2.5mm (0,1") matrix.

BIMBOARD 1 £ 8.83

BIMBOARD 2 £21,01

BIMBOARD 3 £29,84

BIMBOARD 4 £38,79

DESIGNER PROTOTYPING SYSTEM

, 2, or 3 BIMBOARDS mounted on BIM 6007 BIMCONSOLE with Integral Power Supply (±5 to ±15Vdc @ 100mA and fixed +5Vdc @ 1A) All O/P's fully isolated. Short circuit and fast fold back protection. Power rails brought out to cable clamps that accept stripped wire or 4mm plug.

> DESIGNER 1 £55.62 DESIGNER 2 £61.02 DESIGNER 3 £66.42

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MICRO SWIT	CHE	s	C	OIN O	PER	DP	ED 45p DT 25p
TIL 209 RED		1		15	p.	CL	IPS 3p
TIL 209 GREE				***			25p
ENAMELLED	COP	PER	WIRE	32, 34,	36,	38,	
40 SWG	39.	***	***	***			1p FT
RESISTORS		***	1000	***	:00		2p ea.

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Phillips control box for coupling stereo cassette to stereo car radio. Includes stab supply, transformers A.F. 3 Din plugs, 2 Din sockets, 3 P.Bs etc. + diagram £1-46

8" × 5" L.S. units, 4 ohm—8w pk ceramlc magnet. Good hi-power car speakers, pair. £6.00

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Now contains 200 sq. ins. copper clad board, 11b. Ferric Chloride. DALO etch resist pen, abrasive cleaner, two miniature drill bits, etching dish and instructions. £4.25.

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10mm 12V cort Drock at Contacts 20 x 30 x 25mm. Only 56p. W701 6V SPCO 1A contacts 20 x 30 x 25mm. Only 56p. W817 11 pin plug in relay, rated 24V ac, but works well on 6V DC. Contacts 3 pole c/o rated 10A. 95p. W819 12V 125OR DPCO 1A contacts. Size 29 x 22 x 18mm. Min plug-in type 77p.

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2 pin switched speaker socket, PC mounting; 5 pin 180° PC mntg or chassis mntg (clip fix). All the same price, any mix: 10 for 70p, 25 for £1.60, 100 for £5.50.

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.33W	3p	2p	1.76N
UPMO33	•4Ω7 to 1	IM 5% <i>E24</i>	
.33W	2p	1.6	1.43N
CR25	●1 M2 to	10M 10% <i>E1</i>	2
.33W	3р	2p	1.76N
MR25	5Ω1 to	300K 2% E2	4
4W	5p	4p	3.60N
UPM050	•4Ω7 to 4	4M7 5% <i>E12</i>	
.5W	2p	1.6	1 43N
CR37	 1Ω to 30 	295% E12	4 071
.5W	3p	2p	1.87N
UPM075		10M 5% E24	1.43N
.75W		1.6	
UPM100		4M7 5% <i>E12</i> 4p	3.27N
1W UPM100	5p	10M 10% E1	2 3.2,11
1W	5p		3.27N
TR5	100 to	1M 2% E24	
5W	5p	4p	3.20N
TW1	0 220	to $0.47\Omega \pm 0$.05Ω E12
1W	15n	13o	11.8N
TW1	0.56Ω	to 3R9 10%	E12
1W	15o	13p	11.8N
GWS3	0.47Ω	, 1Ω to 10Ω 1	0% <i>E12</i>
3W	18n	14p	11.0N
GWS3	12Ω to	10K 5% E12	?
3W	18p	14p	11.0N
GWS7	1Ω to	10Ω 10% <i>E1</i> 2	?
7W	18p	14p	11.0N
GWS7	12Ω to	10K 5% E12	?

7W 18p 14p 11.0N Net prices apply for complete 100s only. Bulk prices available. E12 values: 1.0, 1.2, 1.5, 1.8, 2.2, 2.7, 3.3, 3.9, 4.7, 5.6, 6.8, 8.2. E24 values: as E12 plus 1.1, 1.3, 1.6, 2.0, 2.4, 3.0, 3.6, 4.3, 5.1, 6.2, 7.5, 9.1, and their

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One-pole change- Two-pole change- Three-pole chang-	42p 78p £1.08		
On-aff 2-pels SDS.2 Da-off 4-pels SDS.4	42p 78p	On-off 6-po Ca-off 8-po	

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£1.08 For fuller range of switches, see section 5 in our current ads.

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100Ω to 4M7 220Ω to 2M2 1K, 4K7 to 2M2 4K7 to 2M2 P20 lin. P20 log: JP20 lin. JP20 log: DP20 made to order in any values available in P20. State front (near bush) and rear tracks clearly.

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travel) For knobs, see JV-slider PG58 mono C1PG58ST stereo Resistance values stocked.

4K7 to 1M 10K to 1M 4k7 to 1M 4K7 to 1M 4K7 to 1M 4K7 to 1M PG58 lin PG58 log: C1PG58 lin C1PG58 log C1PG58ST lin:

C1PG58ST log. ± 2dB @ 10%F ± 3dB @ 3 2% Log stereo matching



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AB8	101	x	10,1	x	38	60p
AB9	101	×	70	X	3 8	60p
AB10	133	х	101	×	38	62p
AB11	101	x	64	×	51	60p
AB12	76	х	51	x	25	50p
AB13	152	х	101	x	51	77p
AB14/2	127	x	89	×	64	74p
DIECAST						
992	89	x	35	x	30	1.24
993	114	x	64	x	30	1.32
998	114	×	64	×	55	1.62
994	114	x	89	×	55	2.10
999	171	×	121	×	55	3.07
974	171	Ξ.	121	×	106	4.39
PLASTIC	116	x	77	x	36	48p
PB1	122	x	67	x	43	66p
PB801	, 22	^	0,	^	, -	

(PB301 is "double U" type and has vent and various holes!

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 IN UK OVER 25 LIST VALUE. If under, add 27p
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Knobs

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EV18A 18mm				٠.	56p
EV22A 22mm					64p
EV32A 32mm					73p
EV38A 35mm	1				78p

K107 pointer 32mm black or



21 LWM COFFEE WAS	
(Caps must be ordered \$	eparately
15mm diameter	
S150B short	27
K 150B	28
W1508B winged	32
22mm diameter	
S210B short	35
K210B	36
W210B winged	39
Caps 15mm	
C150	3
C151 with line	5
C152 w. spot	5
Caps 21mm	_
	3
C210	

5p C212 w. spot Nut covers 15mm only: N150B N151B with line Dials 15mm

Dials 15mm
Oials 21mm
Types available
D151, D211 0-11
D152, D212, 1-12
D158, D218 wedge
D159, D219 white arrow

Caps in black, red, yellow, green, blue and grey.
All other items black only

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Plastic cylindrical — black, red, yellow, green, blue, light or dark grey, white JV18 18mm JV18 18mm 27p JV23 23mm 27p The following pair are designed for DP20 pots, and have 4mm and 6mm bores 40p 40p JV18/4 JV23/6T

JV Slider knobs for Radiohm mono and stereo pots. In eight 8р

colours as above	
JVS + colour	
BLACK PLASTIC	
K1 25mm S	13
CK1 25mm SM	18
K2 35mm S	15
CK2 35mm SM	26
K3 27mm P	12
K4 32mm P	12
K5 19mm	20
K6 25mm	29
NK 36mm SK	3
PK 36mm SP	3
SK6 41mm SK	30
K7 19mm B/W	20
K8 25mm S	20
S skirted	
M metal insert	
P pointer	
Sk skirt, 0-10	
B/W black or white	











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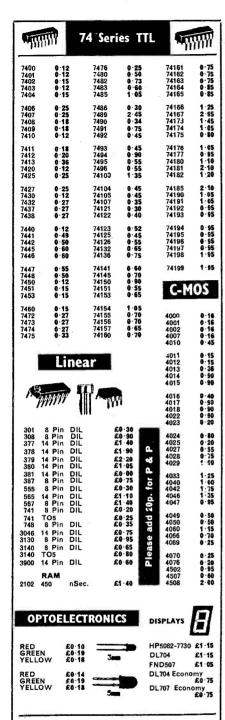




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TEMPUS

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international

roduction of the new catalogue has been held up for a few weeks - since we have just been appointed as distributors for two of the most exciting ranges of radio components products yet: The Micrometals range of iron dust torroids cores and formers, and the OKI range of VLSI for digital frequency displays for receivers. We apologize for any inconvenience, but these two ranges are really worth the wait, and include some products you will find hard to believe, like the MSM5523 IC, an IC with less than ten external components that gives AM frequency readout to 1kHz from LW to 39.999MHz, FM frequency readout in 100kHz steps - (all usual IF offsets programmable by diodes), a 24 hour format clock with 12 hour display, independent on and off timers, time signals on the hours, stopwatch facility and a sleep timer. This costs £14 with its timebase crystal, and makes all that has gone before an expensive and time wasting excercise. Rather like the way the Intersil ICM7216 has revolutionized the instrument counter market. (See the OSTS ad.) And those of you familiar with Amidon and IG dust torroids, favoured in many new RF designs, will be pleased to know Ambit will be stocking a broad range of the Micrometals types for applications from EMI filters to RF PA stages.

OKI frequency counter ICs: details in cat2

for CA LEDs with RHDP such as FND507 £14 inc xtal for 3½ digit LCD AM/FM with MSM5525 direct segment drive, no clock or timers £11 inc xtal Other types for fluorescent displays etc OA

emiconductor additions: pilot cancel mpx decoder muting stereo preamp supercedes TDA2020 HiFi AM/FM low cost AM/FM KB4438 HA1370

TDA1090 1 45 TDA1220 PRICES DOWN ON VMOS: as expected, this new technology in power transistors is getting cheaper, 120v comp pairs /100W for £10.00 Price reduction on CA3189Enow £2.20
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Some resultanz for RF portifically: 8F256LB/0,34; 40822/0,43*,40823/0.51 * 40873/0.55* BF900/961/0.80*,8F960/1.60* 8F224/0.22; BF274/0,18; BF195/0,18; BF240/0.22; BF241/0.22; BF362/0,70; BF479/0.86; BF695/0.70; BFY90/0.90*

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At last, DIV Hi Fi which looks as if it isn't.

That's not to say it doesn't look like HiFi - just that it doesn't look like the usual sort of thing you have come to associate with DIY HiFi. The Mk3 outstrips and outperforms all British made HiF! tuners, and most imported ones too. Certainly at the price, there isn't one near it. But more than that, it looks superb. A small pic here would be an insult, so send an SAE for details on the kit that looks as if isn't. It's something else......

- Exceptionally high performance exceptionally straightforward asset
- Exceptionally night performance exceptionally straightforward assembly Baseboard and plug-in construction. Future circuit developments will readily plug in, to keep the MkIII at the forefront of technical achievement Various options and module line-ups possible to enable an installment approach

and now previewing the matching 60W/channel VMOS amplifier:

- Matching both the style and design concepts of the MkIII HiFi FM tuner Hitachi VMOS power fets characterized especially for HiFi applications Power output readily multiplied by the addition of further MOSFETs
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The PW Dorchester·LW.MW.SW.& FM stereo tuner



In much the same way as we have swept away the 'old technology' in frequency/timer radio tuner. Don't confuse this one chip radio with things like the ZN414 · for this is a genuine superhet receiver with a mechanical AM IF filter, and ceramic IF filters for FM. genuine superhet receiver with a mechanical Aim is fitter, and ceramic if it fitters for Fin.

The AM section employs a balanced input mixer section, covering all broadcast bands - plus a BFO and MOSFET product decetor for SSB/CW - though at this price, the tuner is not intended as a "communications receiver" - although we know of many lesser designs that make that claim. The AM sensitivity is nevertheless better than 5uV, and FM sensitivity is 1.2uV for 30dB S/N. As a multiband broadcast superhet receiver, it is a unique constructor project that fulfills the requests we very frequently get for a general coverage circuit that sin't over complicated. The set has CA3089E FM performance, with mute etc., and a PLL stereo decoder with full pilot tone filtering.

The tuner board - with "on board" PCB mounted switching, all components etc : £33.00 The case/cabinet with PSU, meter and mechanics etc

An SAE for full details please. See the feature article in Practical Wireless (Dec/Jan)

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4012	17p	4072	20p	4538	150p
4013	55p	4073	20p	4539	110p
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4017	80p	4076	90p	4543	174p
4018	80p	4077	20p	4549	399p
4019	60p	4078	20 p	4553	440p
4020	93p	4081	20p	4554	153p
4021	82p	4082	20p	4556	77p
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7401	13	20	7460	17	1 1	74128	74	1	74188	275	1	74378	93
7402	14	20	7463			74132		78	74190	115	92	74379	130
7403	14	20	7470	28		74133		29	74191			74386	37
7404	14	24	7472	28	1 1	74136		40	74192	105	180	74390	140
7405	18	26	7473.	32	1.1	74138		60	74193	105	180	74395	139
7406	38		7474	27	38	74139		60	74194	105	187	74396	133
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7409	17	24	747B		1	74143			74198	150	110	74445	90
7410	15	24	7480	48	1	74144 74145		i 1	74199	160]	74490	140
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7427	27	29	7496	58	120	74161		78	74279		52	LCD D	
7428	36	32	7497	185	1	74162		130	74283		120		2480p
7430	17	24	74100	119	1	74163		78	74290		90	3½ digit	
7432	25	24	74104	63	1	74164		1	74293	-	95	display	1150p
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7437	40	24	74107	32	38	74166		i	74298		100		t 2065p
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7445	94		74116 74118	198 83	1	74175		110	74362		715	divide b	
7446	94	Į			1	74176		1	74365		49		420p
7447	82	100	74119	119	1	74177		1	74366		49	9 5 H900	C 780p
7448	56	99	74120	115	i	74180		1	74367	•	43		C 1400p
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7450 7451	17 17	24	74122	48	1	74182		300	74373		77	by 100	
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7454	17	24		38	1 44	74184		1210	74375		60	i	450p
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38 | 7454 17 | 24 | 74125 38 | 44 | 74184 135 | I

The ICL7216BIPI is still the cheapest way to make a full 8 digit/ 10MHz frequency counter/timer, and with 10 external components + display - It is also one of the simplest. For £19.82, it takes a lot of beating. The males fitters have been extended now to include a 6amp IEC version at £5.10, and with the amount of electronic noise on the average supply (next door's fridge, for intance) it is a really worthwhile addition to any sensitive equipment. LPSN TTL now includes many more of lacest types, all - of course - are absolutely prime first quelity types. And don't forget our range of OPTO displays includes Hewlett Package high efficiency 0.43" (types in all colours - renowned as the finest quality in the market. For other types of component - discrete LEDs, radio and audio devices, tuner modules, kits etc., see our other advertisement for more details - or send for the AMBIT catalogue system. Part one (45p) includes details of our background 'standard' tens, and the new part two includes all the latest introductions and developments, plus a rundown on OSTS.

95p 23p 91p 69p 51p

We have hidden the most important part

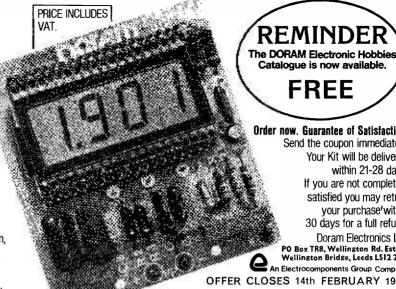
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(BLOCK LETTERS)

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For the advanced, keen short wave listener, the choice of receiver has usually been between cheap and nasty or very good but very expensive equipment. We think that the SRX-30 will provide that listener with excellent performance at a reasonable cost and is the answer to this eternal problem.

able cost and is the answer to this eternal problem.

The SRX-30 provides AM, CW, USB and LSB reception on all frequencies from 500 kHz to 30MHz. All right, so does your Sooper Blooper Mk. 3 but you can't set the Sooper Blooper dial to the frequency you want and be sure that it's correct! The SRX-30 tuning system is so simple to operate. You have a dial reading in MHz from 0-29 and a main tuning dial reading 0-1000 kHz. So—if you know that Radio Slobovia is broadcasting on 10.295 MHz, you set the MHz dial to 10 kHz dial to 10 kHz dial to 295 and there you are. The MHz dial setting is not critical, as stability is guaranteed by a triple mixing drift cancelling system, thereby overcoming another problem in your Sooper Blooper Mk. 3; drift.

another problem in your Sooper Blooper Ms. 3; dritt.

A further drawback to cheap receivers is massive image interference on the higher frequencies due to the use of a low IF, typically 455 kHz. The cure for this problem is the use of a high IF and the SRX-30 employs a first IF of around 40 MHz—so goodby to first IF images. You could of course find the same system as this in the Racal RA17 series receivers: after all, the SRX-30 has copied the basic idea from this very receiver. The big drawback to the RA17 (apart from the price!!) is that unless you have the muscles of a prize fighter, lifting the RA17 may send you for a holiday at Hernia Bay (staying at the Truss House?).

To summarize, the SRX-30 covers 500 kHz to 30 MHz with excellent dial readout and reset accuracy; it has all mode (AM, CW, SSB) reception and is equally at home in broadcast or amateur bands; it has all the facilities of a top class communications receiver, RF gain, fine tuning, selectable sidebands, built in loud-speaker, operation from ac mains or 12v, Dc, rugged construction and super styling and all at an attractive price—£175 inc. VAT. Carr £3.

See it soon at your nearest stockist, you will be agreeably impressed.

For all that's good in Amateur Radio, contact: LOWE ELECTRONICS LTD., 119 Cavendish Road, Matlock, Derbyshire. Tel: 0629 2430 or 2817.

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BC238/A/B BC238C BC328 BC337 BC338 BC516/7	.10 .12 .11 .11 .11	.081 .164 .091 .091 .095	.069 .088 .077 .077 .08	TIP35B TIP36A TIP36B TIP41A TIP41B TIP41C	1.50 1.48 1.63 .54 .79	1.35 1.33 1.46 .48 .714 .575	1,138 1,133 1,235 ,405 ,602 ,485	2N5172 2N5245 2N5298 2N5458 2N5460 3N201 RCA40347	.10 .33 .48 .10 .39 .58	.091 .298 .433 .091 .35 .52	.877 .251 .365 .077 .291 .438 .698	7425 7426 7430 7432 7437/8 7440 7441A	21 .16 .18 .20 .16	.182 19 .144 .155 .18 .344 .665	.154 .18 .122 .131 .152 .122	NE556 SFC2741 SN76110 SN76023 SN76227 SN76560 TAA621A	.57 .25 .73 1.73 .84 .51	.513 .228 .657 1.55 .845 .458 2.47	.452 .192 .554 1.31 .712 .384 2.00	LEDs TIL 209 TIL 212-1 TIL 216-1 TIL 228 TIL 228	.17 .21 .21 .16	.152 .186 .186 .14	.128 .156 .156 .118 .174	1.15 .22 Vantalum Cap 1, 22, 47, 35V	.34 .3 .43 .3	35 19 87 25 89 32
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BCY58/9 BCY70/1/2 BCY78 BD234 BF224 BF241	.17 .18 .16 .57 .18	.148 .171 .133 .509 .14	.125 .144 .105 .429 .118 .118	2N1131 2N1132 2N1613 2N1711 2N1893 2N2218	.26 .30 .21 .27 .36 .34	.235 .269 .19 .237 .321 .304	.198 .227 .16 .20 .27 .256	RCA40412 RCA40673 RCA40841 RCA40871 RCA40872	.68 1.00 .77 .94 1.06	.613 .902 .689 .849	.517 76 .581 .716 .805	7473/4 7478 7489 7481/2 7483 7484 7486/90	.26 .31 .36 .57 .31 .45	.228 .273 .321 .513 .273 .40	.192 23 27 .432 23 .338	UA747 UA748 SN15332	.80 .42 .25	.716 .376 .228	.603 .316 .18	1N4002 1N4003 1N4004 1N4005 1N4006 IN4007	.05 .05 .055 .055 .066	.84 .843 .846 .848 .051 .865	.036 .037 .039 .04 .044 .044	33 10V, 47 6.3V 100 3V 10 6.3V 10 16V 10 35V 22 6-3V	.11 .0 .14 .1 .17 .1 .12 .1	33 .1 195 .0 19 .1 53 .1 06 .0
8F244B 8F257 8F258 8F337 8FR39 8FR40 8FR41	21 .30 .31 .32 .16 .16	.188 .273 .285 .285 .146 .142	.158 .23 .24 .24 .123 .12	2N 22 18A 2N 22 19 2N 22 19A 2N 22 21 2N 22 21A 2N 22 22 2N 22 22	.28 .39 .27 .21 .18	.266 .25 .347 .237 .182 .159	.224 .211 .293 .20 .154 .134	4000A 4001A 4011/12A 4013B 4014A 4017A	.19 .19 .19 .45 .95	.171 .171 .171 .399 .855 .931	.144 .144 .336 .72 .784	7491A 7492 7493 7494 7495 7496	.20 .31 .20 .40 .23 .34	.178 .275 .178 .355 .205 .307	.15 232 15 .299 .173 .259	DIL Sockets 8 pin 14 pin 16 pin 18 pin	.13 .14 .16	112 123 138	.094 .104 .116	IN4148 IN4149 Zeners BZY 88C Series	.02	.016 .816	.013 .014	22 16V 22 16V Resistors Carbon Film 5% E12	.17 .1	23 .1 53 .1
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80185 80108 80205 80208 81509 AJE340 APSA05/6	1.68	1.13 1.44 1.21 1.44 .36 .703	.955 1.219 1.018 1.219 .32 .555	2N2907 2N2907A 2N3053 2N3055 2N3702/3/4 2N3705/7/9 2N3706/8	.18 .21 .25 .90 .09 .10	.169 .19 .222 .811 .079 .083	.134 .16 .187 .683 .087 .07	4030A 4040B 4042B 4043B 4044B 4050B 4051B	.45 1.18 .86 .80 .80 .44 1.24	.399 1.06 .779 .722 .722 .399	.336 .896 .656 .608 .808	74150 74151/3 74156 74157 74161/3/4 74167	.45 .29 .36 .29 .36 1.26	.40 .258 .321 .258 .321 1,14	.338 .218 .27 .218 .27 .953	Voltage R				63V 1.5, 6.8. 2.2, 3.3, 4.7 10uf, 15uf 22uf 47uf	.12 .12 .12 .14	.106 .107 .121 .147	.089 .09 .102 .124	VAT 8%. 25p P. £10. Alf other o invoiced, to PC rate	& P. on o rders carrie s.	- Hoders und age will i
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		TEXAS	400	74221 74251	160 p		140p			LINEAR	I.C.s	1	400	TRANS				241 C	78p	2N3866	90p	DIODES	
7400 7401	13p 14p		180p 130p	74259	250p	74LS195	140p	74C161	155p	*AY1-1313	668p	*MC1496 *MC3340	100p 120p	AC127/8 AD149	20p 70p	BFY51/2 22 BFY56 33	B TIF	242A 242C	70p 82p	*2N3903 *2N3905	/4 18p	*BY127 *OA47	12p 9p
7402	14p	74104	65p	74265 74278	90 p 290 p			74C162 74C163		*AY1-5050	212p	*MC3360	120p	AD161/2	45p	BFY90 96	n TIF	2955	78p	*2N4036	65p	*OA81	15p
7403 7404	14p 17p	74105 74107	65p 34p	74279	140p	74LS240				*AY5-131	600p	*MFC4000 MK50398	5 120p 750p	BC107/8	11p	BLY83 700		93055 S43	70p	*2N4058		*OA85	15p
7405	18p	74109	55p	74283	190p	74LS241	175p	74C173		*AY5-1320	320p	NE531	130p	BC109 *BC147/8	11p	BRY39 45 BSX19/20 20	n ort	S93	34p 30p	*2N4060 *2N4061	/2 18p	*OA90 *OA91	9p 9p
7406 7407	32p 32p	74110 74111	55 p 55 p 70 p	74284 74285	400 p 400 p	74LS242 74LS243	175p 175p	74C174 74C175	160p 210p	*CA5019 *CA5046	80p	*NE540 NE543K	200p 225p	*BC149	10p	*BU105 190	p *Z	TX108	12p	*2N4123	4 220	*OA95	9p
7408	19p	74116	200 n	14200	150p	74LS245	175p	74C192	150p	*CA3048	225p	NE555	25p	*BC157/8 *BC159		*BU108 250 *BU205 220		TX300 TX500	13p 15p	*2N4125 *2N4289	/6 22p 20p	*OA200 *OA202	9p 10p
7409 7410	19p 15p		130p 210p	74293 74294	150p 200p	74LS251 74LS257		74C193 74C194	150p 220p	CA3080E	72p	NE556	70p	*BC169C	11 p 12 p	*BU208 240	p *Z1	TX502	18p	*2N4401	/3 27p	*1N914	4p
7411	24p		210p	74298	200p	74L\$259	175p	74C195	110p	*CA3089E *CA3090A		NE561B NE562B	425p 425p	*BC172	12p	*BU406 145 MJ481 175		FX504 457A	30p 250p	2N4427 *2N4871	90p 60p	*1N916 *1N4148	7p
7412	20p		28p	74365 74366	150p 150p	74LS298 74LS373		74C221	175p		375p	NE565	130p	BC177/8 BC179	17p 18p	MJ491 200			250p	*2N5087	27p	1N4001/2	4p 5p
7413 7414	30p 60p		48p 55p	74367	150p	74LS374	195p	4000 S	ERIES 15p	CA3130S CA3140E	100p 70p	NE566 NE567	155p 175p	*BC182/3	10p	MJ2501 225	D 2N6	597	25p	*2N5089 *2N5172	27p	1N4003/4	6p
7416	27p	74125	55p	74368	150p	81LS95 81LS96	120p 160p	4001	17p	CA3160E	75p	RC4151	400p	*BC184 BC187	11 p 30 p	MJ2955 100 MJ3001 225		597 706A	45p 20p	2N5179	27p	1N4005 1N4006/7	6p 7p
7417 7420	27p 17p	74126 74128	60p	74390 74393	200p 200p	81LS97	120p	4002 4006	17p	FX209 ICL7106	750p 925p	*SN76003N	175p	*BC212/3		*MJE340 65	D 2N7	708A	20p	2N5191	83p	1N5401/3	14p
7421	40p	74132	75p	74490	225p	81L S 98 8T28	160p	4007	95p 18p	ICL8038	340p	*SN76013N *SN76013N	140p	*BC214	12p	MJE2955 100 MJE3055 70	p 2N9		45p	2N5194 *2N5245	90p 40p	1N5404/7	
7422 7423	22p 34p		75p 70p	74 LS SERIES	2	9301	230p	4008	80p	LM301A	36p	F	120p	BC461 BC477/8	38p 30p	*MPF102 45	p 2N1	131/2	20p	*2N5296	55p	2 7V-33V	<i>_</i>
7425	30p	74142	290p	74LS00	18p	9302	175p	4009 4010	40p 50p	LM311 LM318	190p	*SN76023N *SN76023N		*BC516/7	50p	*MPF103/440 *MPF105/640	p 2N1 p 2N1		25p 25p	*2N5401 *2N5457/	50p 8 40p	400 mW	9p 15p
7426 74 27	40p 34p	74145 74147	90p	74LS02 74LS04	18p	9308 9310	316p 275p	4011	17p	LM324	70p		120n	*BC547B *BC549C	16p 18p	*MPSA06 30	p 2N2	102	60p	*2N5459	40p	SPECIA	
7428	36p	74148	150p	74LS08	20 p 22 p	9311	275p	4012 4013	18p 50p	LM339 LM348	90p 95p	*SN76033N *SP8515	175p 750p	*BC557B	16p	*MPSA12 50 *MPSA56 32	p 2N2	160 219A	120p 20p	*2N5460 *2N5485	40p	OFFERS	•
7430 7432	17p 30p	74150 74151	100p	74LS10	29p	9312 9314	160p 165p	4014	84p	*LM377	175p	*TBA641B	11	*BC559C	18p	I "MPSU06 63	p 2N2	222A	20p	*2N6027	48p	100+ 741 £16	
7433	40 p	74153	70p	74LS13 74LS14	38p 100p	9316	225p	4015 4016	84p 45p	*LM380 *LM381 A N	75p	*TBA800	225p	BCY70 BCY71/2	18p 22p	*MPSU56 78 OC28 130		369A	16p 30p	2N6247 2N6254	190p	100+ 555	,
7437 7438	35p 35p	74154 74155	100p 90p	74LS20 74LS22	22p 28p	9322 9368	150p 200p	4017	80p	*LM389N	150p	*TBA810	100p	BD131/2	50p	OC35 138	p 2N2	646	50p	2N6290	130p 65p	£20 100+	
7440	17p	74156	90p	74LS22	28p	9370	200p	4018 4019	89p 45p	LM709	36p	*TBA820	90p	BDY56 BF200	200p 32p	*R2008B 200 *R2010B 200		904/5 906A	25p	2N6292	65 p	RCA 2N3	3055
7441 7442A	76p	74157 74159	70p	74LS30 74LS47	22p	9374 9601	200p	4020	100p	LM710 LM733	50p 100p	*TCA940 *TDA1022	175p	*BF244B	35p	*TIP29A 40	p 1 2N2	907A	24p 30p	2N128 3N140	120p 100p	£36 BRIDGE	
7443	112p	74160	190p 100p	74LS55	90p 30p	9602	175p	4021 4022	110p 100p	LM741	29p	XR2206	400p	*BF256B BF257/8	78p 32p	*TIP29C 55		2926	9p	3N201	110p	RECTIF	ERS
7444 7445	112p	74161 74162	100p 100p	74LS73 74LS74	50p	9603 INTERF	60p	4023	22p	LM747 LM748	70p 35p	XR2207 *XR2216	400p 675p	BF259	36p	*TIP30C 60i		053 054	20p 65p	3N204 40290	100p 250p	*1A 50V *1A 100V	21 p 22 p
7446A	93p	74163	100p	74LS75	40p 50p	I.C.s		4024 4025	50p 20p	LM3800	70p	XR2240	400p	*BFR39 *BFR40	30p	TIP31A 58			48p	40360	40p	*1A 400V	30p
7447A 7448	70p 80p	74164 74165	120p	74LS83	110p	MC1488 MC1489	100p	4026	130p	LM3911 ·	130 p	*ZN414 ZN424E	90p 135p	*BFR41	30p	TIP32A 68	2N34	553	140p 240p	40361/2 40364	45p 120p	*2A 50V *2A 100V	30p 35p
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7451 7453	17p	74167	200p	74LS90	60p	75182	230p	4029	100p	MC1458	55p	ZN1034E	200p	*BFR81	30 p i	TIP33A 90; TIP33C 114		3643/4 3702/3		40409 40410	65p	*3A 200V *3A 600V	60p 72p
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7482 7483 A	84p	74184A	150p	74LS158	120p	74C30	27p	4047	110p	100mA 7 5V 78L05	O-92 35p	100mA 7 5V 79L0b	O-92 80p			w	hich ;	are a	at 12	1%			- 1
7484	99p 100p		150p 700p	74LS160 74LS161	100p	74C32 74C42		4048 4049	55p	12V 78L12	35p	12V 79L12	80p		_		_	_	_	_			
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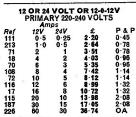
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CA3014* 137	MC1310P
CA3018* 68	MC1312PO
CA3020 176	MC1312PQ MC1488*
CA3023 176	MC1489*
CA3028A* 80	MC1489* MC1495
CA3035 240	MC1496L MC3340P* MC3360P MC3401
CA3036 110	MC3340P*
CA3043 199	MC3360B
CA3045 140	MC3401
CA3046 70	MEM780
CA3048 200	MFC4000B
CA3075 175	MFC6040*
CA3080E* 70	MK50362*
CA3081 190	MK50398*
CA3089E 210	MM57160*
CA3090AQ 375	
CASUSUAU 313	NE543K
CA3123 200	NE544
CA3130* 85	NE555*
CA3140 70	NE556DB*
ICL7106* 752*	NE560*
ICL7107* 975	NE561*
ICL8038CC+ 335	NE562B*
ICM7205* 1150	NE564*
LD130* 452	NE565A*
,00	

195	ROM2513*	595	7403	14	7481	- 33
70	SG3402*	295	7404			69
80	SL437A	560	7405	18	7482	
	SN72710*	43	7406	30	7483	72
95	SN72733*	125	7407	38	7484	95
125			7408	17	7485	106
375	SN76003N	175	7409	17	7486	31
95	SN76013N	140	7410	15	7489	210
145	SN76013ND	120	7411	20	7490	33
248	SN76018*	148	7412	17	7491	75
125	5 N76023N	146	7413	30	7492	38
125	SN76023ND	111	7414	51	7493	32
50	SN76033N	175	7416	30	7494	78
60	SN76115N	215	7417	30	7495	65
76	SN76131*	110	7420	16	7496	57
125	SN76227N	116	7421	29	7497	189
750	SN76477*	225	7422	24	74100	119
785	TAA621AX1	228	7423	27	74104	62
88	TAA661A	155	7425	27	74105	82
260	TAA960	300	7426	36	74107	29
149	TAD100	150	7427	27	74109	64
195			7428	35	74110	54
85	TBA120S	70	7430	17	74111	68
96	TBA540	215	7432	25	74112	98
395	TBA550Q	330	7433	40	74116	198
92	TBA641-A12		7437	39	74118	83
150	BX1 or BX11	250	7438	33		119
120	TBA651	180	7440	17	74120	115
70	TBA800	90	7441	74	74121	25
205	TBA810S	99	7442	68	74122	46
85	TBA820	70	7443	115	74123	48
97	TBA9200	260	7444	112	74125	38
650	TCA965*	120	7445	94	74126	57
635	TDA1022	575	7446	94	74128	74
620	TDA2020	320	7447	82	74132	73
210			7448	56	74141	56
185	TLO71*	70	7450	17	74142	269
29	TLO81 CP*	52	7451	17	74143	314
60	TLO82CP*	96	7453	17	74144	314
325	TLO84CP*	130	7454	17	74145	65
395	UAA170	198	7460	17	74147	175
410	ZN414	90	7470	28		109
425	ZN424E	130	7472		74150	99
120	ZN425E*	410	7473	39	74151	64
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170 NE566* 39 NE567V* 240 NE571 110 RAM2102 120 RC4138D

87 74160 82 96 116 4018 89 4072 22 4511 74160 92 107 44 4019 48 4073 22 4511 82 74162 92 132 95 4020 99 4075 23 4513 74164 105 151 96 4022 88 4077 40 4515 74165 105 151 96 4022 88 4077 40 4515 74165 105 151 96 4022 10 4076 82 4516 74166 101 155 96 4024 85 4081 20 4517 74167 102 30 157 78 4028 100 4085 21 4517 74177 230 157 78 4028 100 4085 21 4517 74177 230 157 78 4028 100 4085 21 4517 74172 230 157 78 4028 100 4085 21 4517 74172 230 157 78 4028 100 4085 21 4517 74172 102 31 158 94 4027 45 4086 73 4529 174175 107 160 122 4028 11 4089 150 4521 74175 107 160 123 4028 11 4089 150 4521 74175 107 160 123 4028 11 4089 150 4521 74175 107 160 123 4028 11 4089 150 4521 74175 107 160 123 4028 11 4089 150 4521 74175 10 4033 13 4098 110 4527 84174 100 453 113 4032 10 4097 172 4529 174 185 185 181 134 4032 100 4097 172 4529 174 185 185 181 134 4032 100 4097 172 4529 174 185 185 181 134 4032 100 4097 172 4529 174 190 155 193 130 4004 100 4534 4165 100 4534 4165 100 4534 4165 100 4538 4174 100 4534 4165 100 4538 4174 100 4538 100 4041 100 4534 4175 100		74160	82	90	•••	70	-	40/1	21	4010	
74101 2 2 107 44 4019 44 4073 21 4512 74102 2 1076 1076 2 2 1512 74103 82 139 5 4020 99 4075 23 4512 74103 82 139 85 4021 91 4076 23 4512 74105 105 153 76 4022 83 4077 40 4515 74105 105 153 76 4022 80 4076 21 4516 74105 105 153 76 4022 80 4076 21 4516 74107 198 156 96 4022 1 4076 81 20 4517 74107 198 156 96 4025 19 4082 21 4518 74172 923 157 76 4022 190 4085 74 4518 74173 77 160 162 4022 81 4086 73 4520 174173 77 160 161 98 4022 81 4086 73 4520 174174 87 161 98 4022 99 4093 85 4522 174175 77 162 162 133 4030 99 4093 85 4522 174176 77 161 98 4029 99 4093 85 4522 174176 77 162 133 4031 285 4096 190 4057 74173 74177 77 17 163 113 4031 285 4096 190 4528 74 4181 885 164 114 4033 143 4098 115 4530 174184 235 181 318 4035 111 4150 190 4534 174188 275 191 140 4037 190 4162 190 4534 174191 105 193 130 4038 22 4174 110 10 4534 174191 105 193 130 4038 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 241 236 4042 79 4408 728 4543 174191 105 193 130 4038 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4534 174191 105 193 130 4039 22 4174 110 4539 174191 105 193 130 4039 22 4174 110 4539 174191 105 193 130 4039 22 4174 110 4539 174191 105 193 130 4039 22 4174 110 4539 174191 105 193 130 4039 22 4174 110 4539 174191 105 193 130 4039 22 4174 110 4539 174191 105 1452 100 4036 184 4109 278 4555 174191 105 193 130 4088 184 4409 728 4555 174191 105 193 130 4088 184 4409 728 4555 174191 105 193 130 4088 184 4409 728 4555 174191 105 193 130 4088 184 4409 728 4555 174191 105 193 130 4088 184 4409 728 4555 174191 105 193 4009 184 4419 728 4555 174191 105 193 130 4088 184 4409 728 4555 174191 105 193 1404 4408 728 4449 728 4455 128 4449 728 4455 128	ч			96.	116			4072	21	4511	
74102 #2 132 95 4020 99 4076 23 4513 74104 103 151 96 4022 88 4077 40 4515 74105 103 151 96 4022 88 4077 40 4515 74105 103 155 96 4022 10 4078 21 4516 74106 108 155 96 4022 10 4085 21 4518 74170 230 156 96 4022 10 4082 21 4518 74170 230 156 96 4022 10 4085 74 4519 74170 283 157 76 4022 10 4085 74 4519 74171 70 158 96 4027 45 4086 73 4529 74173 77 161 98 4022 81 4089 150 4521 74174 87 160 122 4022 81 4089 150 4521 74176 77 161 98 4022 99 4093 85 4521 74178 77 161 98 4022 99 4093 85 4521 74180 85 164 114 4033 143 4096 105 4521 74180 85 174 106 4034 4097 108 4096 105 4528 74180 85 176 106 4035 117 4099 116 4527 74181 99 175 110 4035 111 4099 110 4527 74181 99 176 106 4037 109 4150 109 4150 74180 115 192 130 4039 229 4165 109 4537 74180 150 193 130 4040 168 4177 109 4538 74180 150 193 130 4040 168 4177 109 4538 74180 150 193 130 4040 168 4177 109 4538 74180 150 193 130 4040 168 4177 109 4538 74180 151 195 109 4041 80 4177 109 4538 74180 151 195 109 4041 80 4177 109 4538 74180 151 195 109 4041 80 4177 109 4538 74180 151 195 109 4041 80 4177 109 4538 74180 151 195 109 4041 80 4177 109 4538 74180 151 195 109 4041 80 4177 109 4538 74180 151 140 4035 111 4539 74180 151 151 140 4035 111 4059 110 4538 74180 151 151 140 4038 110 4059 110 4539 74180 151 151 140 4038 110 4037 110 4538 74180 151 151 140 4038 110 4039 110 4539 74180 151 151 151 151 151 151 151 151 151 15	,			107	44					4512	
74195 105 151 56 4022 22 4076 23 4514 74195 105 153 76 4023 22 4076 23 4516 74195 25 4514 74195 22 4006 15 4037 122 4516 4023 4076 23 4516 74195 25 4514 74195 22 4006 15 4037 122 4076 23 4076 23 4076 24 408 78 4081 20 4517 4177 178 161 198 4025 198 4082 21 4518 4086 71 4518 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081 150 4818 4081	2					4020	99				
74195 105 153 78 4022 28 4077 21 4515 74105 105 153 78 4023 28 4077 40 4515 74106 101 153 78 4024 69 4077 21 4516 74170 230 158 88 4025 19 4081 20 4517 74172 230 158 88 4025 19 4082 21 4518 74173 170 158 96 4028 1 4085 74 4519 74173 170 158 96 4028 1 4085 74 4519 74173 170 158 96 4028 1 4085 74 4519 74175 87 161 98 4028 1 4085 74 4519 74177 78 162 133 4031 283 4094 190 4521 74177 78 163 113 4032 190 4099 155 4522 74178 181 183 184 4032 190 4097 372 4529 74181 185 174 106 4034 194 4097 372 4529 74182 275 191 140 4038 194 4099 145 4531 74188 275 191 140 4038 194 4099 145 4531 74188 275 191 140 4038 194 4162 190 4538 74191 105 193 130 4040 184 4162 190 4538 74191 105 193 130 4040 184 4162 190 4538 74191 105 221 98 4042 75 4194 418 4531 74192 155 196 100 4041 100 4175 99 4541 74193 105 221 98 4042 75 4194 418 455 257 74197 155 298 185 4044 8 4410 720 4538 74197 155 298 185 4044 8 4410 720 4538 74197 15 228 4000 15 4052 77 4435 825 4560 74197 15 4000 15 4052 77 4435 825 4560 74197 15 4000 15 4052 77 4435 825 4560 74197 15 4000 15 4052 77 44450 255 4565 10 20 4007 18 4055 72 4445 255 4565 10 20 4007 18 4055 72 576 4450 255 4585 13 33 4008 87 4057 275 276 4450 255 4581 14 75 4009 50 4069 115 4490V 525 4581		74163	62			4021	95				
74165 105 151 37 8 4024 28 4078 21 4516 74167 198 155 86 4025 19 4081 21 4518 74167 198 155 86 4025 19 4081 21 4518 74172 230 156 80 4025 19 4082 21 4518 74172 230 157 78 4027 41 4085 71 4519 81 4028 41 4085 71 4519 4028 41 4085 71 4519 4028 41 4085 71 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4088 173 4529 4028 41 4089 110 4528 4029 4029 4029 4029 4029 4029 4029 4029		74184	105	138							
74196 191 153 78 4024 4081 20 4517 4519	í										
74172 923 157 78 4025 190 4082 21 4518 74172 923 157 78 4027 45 4082 21 4518 74172 925 157 78 4027 45 4085 73 4529 74173 170 158 96 4028 81 4086 73 4529 74174 87 160 122 4029 93 81 4086 73 4529 74175 77 167 180 122 4029 93 81 4088 130 4521 74176 75 162 138 4031 25 4094 190 4527 74177 77 16 153 114 4032 25 4094 190 4527 74177 78 153 114 4033 145 4097 137 4529 174180 85 164 114 4033 145 4097 137 4529 174181 135 144 106 4037 140 4098 110 4527 74182 90 175 110 4034 140 4098 110 4527 74182 150 140 4035 110 4037 140 4098 110 4530 74182 150 140 4037 140 4038 141 4098 110 4530 74182 150 150 160 4037 160 4101 160 4534 74182 150 150 160 4037 160 4101 160 4534 74180 151 180 190 150 180 4037 160 4101 160 4534 74197 85 Complete 4043 180 4098 180 4555 74197 85 Complete 2040 150 4037 140 4415 170 4538 74197 85 Complete 2040 150 4037 74 4408 170 4553 74198 150 180 4000 15 4057 72 4438 425 4555 10 20 4007 18 4007 77 4455 225 4555 10 20 4007 18 4005 50 4059 4409 425 4559 114 75 4009 50 4059 4409 55 5581 13 38 4008 87 4057 72 74450 225 4558 13 38 4008 87 4057 72 75 74450 255 4581 14 75 4009 50 4059 110 4450 255 4581 14 75 4009 50 4059 110 4450 255 4581 15 13 38 4008 87 4057 72 75 74450 255 4581 15 13 38 4008 87 4057 72 74450 255 4581 15 13 38 4008 87 4057 72 73 4435 225 4550 14 75 4009 50 4059 110 4450 255 4581				153			20	4078	21		
74170 230 157 76 4022 190 4082 21 4518 74172 823 157 76 4022 190 4085 74 4519 74172 823 157 76 4022 190 4085 74 4519 74172 823 157 76 4022 1100 4085 74 4519 74174 87 160 122 4022 1100 4085 154 4028 73 4520 174 177 78 162 138 4030 38 4094 190 4521 74177 78 163 113 4033 285 4096 190 452 174 181 185 174 106 4033 143 4098 110 4530 174 181 185 174 106 4033 143 4098 110 4530 174 181 185 174 106 4033 143 4098 110 4530 174 181 185 174 106 4037 110 4037 110 4037 110 4037 110 4037 110 4037 110 4037 110 4038 111 4150 110 4534 174 181 181 181 181 181 181 181 181 181 18								4081	20	4517	
74172 923 158 94 4027 45 4086 73 4529 74173 170 160 123 4028 81 4089 150 4521 74173 170 160 123 4028 81 4089 150 4521 74175 87 161 98 4029 99 4093 85 4522 74175 875 162 133 4030 83 4094 190 4027 74176 875 163 113 4031 285 4096 100 4627 74170 78 163 113 4031 285 4096 100 4627 74180 185 174 106 4033 145 4099 107 4180 185 174 106 4033 145 4099 114 4531 74182 135 190 140 4034 180 4099 145 4531 74182 135 190 140 4037 180 4101 190 4536 74180 180 180 180 180 180 180 180 180 180				156	96			4082		4518	
158				157	76	4026					
140	8		925			4027	45		73		
7.417.4 87 161 89 4029 99 303 85 4522 7.4178 87 162 133 4039 88 4034 180 4527 7.4178 75 162 133 4039 185 4054 180 4527 7.4178 75 163 113 4031 295 4056 198 4529 17.4180 85 174 106 4033 144 4059 114 4530 17.4180 185 174 106 4033 144 4069 114 4530 17.4180 185 174 106 4035 111 4150 180 4534 17.4182 195 181 316 4035 111 4150 180 4534 17.4182 135 190 140 4035 111 4150 180 4534 17.4182 135 190 140 4035 181 4150 180 4534 17.4182 185 190 145 180 180 180 180 180 180 180 180 180 180	2					4028	81				
74175 87 102 133 4030 88 4093 89 4327 74177 78 102 133 4031 205 4096 130 4527 74177 78 103 131 4031 205 4096 130 4527 74180 130 140 4033 145 4097 372 4529 4529 175 100 4034 140 4098 145 4531 454 4098 135 181 318 4036 325 4150 140 4034 140 4038 141 410 140 4033 141 410 140 4038 141 410 140 4558 140 4408 140 4408 178 4544 141 441 441 441 441 441 441 441 44	н			1 100							
74178 78 163 143 4031 265 4094 180 4529 74180 85 174 100 4035 114 4032 148 4098 112 4531 74180 85 174 100 4033 144 4098 112 4531 74182 90 181 318 4035 111 4098 112 4531 74182 91 181 318 4035 111 4035 1	5	74175	87								
74177. 78 163 113 4032 4096 4096 108 4528 74181 185 174 108 4034 4097 372 4529 174 181 185 174 108 4034 4097 372 4529 175 174 182 90 175 110 4033 111 4099 114 4531 111 401 4534 111 411 411 411 411 411 411 411 411 4	7	74176	75		138			4094			
74180		74177	78						105		
74181 185 174 106 4032 186 4099 145 4531 74194 135 181 318 4038 325 4161 100 4534 74198 135 191 140 4037 180 4162 190 4538 74191 105 193 130 4038 325 4161 100 4534 74192 185 196 100 4040 180 4175 98 4541 74191 105 193 130 4038 328 4163 100 4538 74191 105 193 130 4038 328 4163 100 4538 74191 105 193 130 4038 328 4163 100 4538 74192 105 193 100 4038 188 4163 100 4538 74192 185 196 100 4040 180 4175 98 4543 74195 95 248 48408 728 4543 74195 95 288 185 4044 88 4408 728 4543 74195 95 288 185 4044 88 4408 728 4543 74195 95 288 185 4044 88 4408 728 4554 74197 85 Complete 187 4198 150 188 4000 187 4194 180 4556 74197 85 Complete 204 405 188 4000 15 4058 145 4410 720 4554 4415 785 4556 100 20 4005 18 4000 15 4058 180 4419 28 4559 4550 4415 4415 785 4556 10 20 4005 18 4005 18 4415 785 4556 10 20 4007 18 4055 72 4440 1275 4556 13 33 4008 87 4057 27 24405 225 4556 13 33 4008 87 4057 27 27 4445 225 4556 13 33 4008 87 4057 27 27 28 4450 25 4556 13 33 4008 87 4057 27 27 28 4450 25 4556 13 33 4008 87 4057 27 27 28 4450 25 4556 13 33 4008 87 4057 27 27 28 4450 25 4556 13 33 4008 87 4057 27 27 28 4450 25 4556 13 33 4008 87 4057 27 27 28 4450 25 4556 13 33 4008 87 4057 27 27 28 4450 25 4556 13 27 28 4010 50 4050 115 4490V 525 4581	1	74180						4097	372	4529	
74182 98 175 110 4033 18 4059 415 4531 74185 135 181 318 4033 325 111 4150 169 4534 74185 135 181 318 4033 325 4161 100 4534 74185 135 181 310 4038 188 4162 100 4534 74180 115 192 130 4038 188 4162 100 4534 74190 115 192 130 4039 220 4174 110 4538 74192 185 196 100 4041 80 4194 100 4538 74192 185 196 100 4041 80 4194 100 4534 74180 105 221 98 4042 75 4408 72 4543 74186 105 241 238 4043 84 4409 720 4554 74186 155 128 1044 88 4419 104 4554 1454 1450 145 145 145 145 145 145 145 145 145 145	1			174	106			4098		4530	
74184 435 181 318 4183 131 4150 169 4532 74185 135 190 46 4037 100 4162 101 100 4534 174188 275 191 140 4037 100 4162 101 100 4534 174190 115 192 100 4038 188 4163 100 4538 174191 105 193 100 4039 220 4174 110 4539 174192 185 196 100 4041 188 4175 90 4539 174193 105 221 90 4041 188 4175 90 4539 174195 95 228 185 4043 8 4409 720 4539 174196 95 298 185 4043 8 4409 720 4534 174197 85 Complete 18 4719 18 47	: 1			175	110					4531	
74185 335 190 140 4038 323 4101 100 4534 74185 74188 275 191 140 4038 188 4165 100 4538 74190 115 192 130 4039 322 4174 110 4538 74191 105 123 130 4039 322 4174 110 4538 74191 105 123 130 4039 322 4174 110 4538 74191 105 123 130 4040 188 4175 100 4538 74191 105 123 4040 188 4175 100 4538 74191 105 123 4040 188 4175 100 4538 74191 105 123 4040 105 1438 4409 720 4554 1451 100 455	: 1			101	240						
74198 275 191 140 403 100 4162 100 4538 74191 105 192 100 4036 108 4165 100 4538 74191 105 193 100 4038 222 4174 110 4539 74192 105 193 100 4041 108 4175 90 4541 74193 105 221 90 4041 75 409 720 4553 74195 95 298 185 4045 409 720 4554 74195 95 298 185 4045 409 720 4554 74197 85 Complete 74198 150 20 4005 100 4041 400 100 4041 400 720 4554 74197 85 Complete 1045 1045 1045 1045 1045 1045 1045 1045	4										
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74191 105 193 120 1040 25 4174 110 4539 74192 125 196 104 104 105 124 124 104 105 124 104 105 124 104 105 124 104 105 124 104 105 124 104 105 124 104 105 124 104 105 124 104 105 124 105 125 105 125 105 105 105 125 125 105 105 105 105 105 105 105 105 105 10	Н					4038	108				
74191 105 193 130 4040 188 31.7 19 4337 74192 185 195 100 4041 80 4175 99 4541 74193 105 221 98 4042 75 4194 408 728 4549 74195 95 298 185 4044 88 4408 728 4549 74195 95 298 185 4044 88 4410 728 4554 74190 95 20 4005 14 410 728 4554 74197 85 Complete 4046 128 4411 728 4554 74198 185 145 4410 728 4556 74197 87 44157 97 44152 138 4557 4558 145 4415 785 4415 785 4	9					4039	329		300		
74192 185 196 100 4041 88 41194 188 4543 74194 105 221 98 4042 75 4198 188 4543 74196 99 386 86 4043 84 4409 720 4554 74198 150 25 281 85 4044 88 4409 720 4554 74198 150 25 281 85 4044 88 4410 720 4554 74198 150 25 Complete 4047 97 4415F 785 4558 4045 145 4119 280 4555 415 4000 18 4000 17 4053 72 4433 1099 4551 04 22 04005 195 4050 72 4433 1099 4551 04 22 04005 195 4050 72 4433 1099 4551 04 22 04005 195 4050 72 4450 255 4562 05 10 20 4007 18 4050 72 4450 255 4562 10 21 475 4009 59 4059 480 44990 595 4580 14 75 4009 59 4059 480 44990 595 4581	ч					4040	105				
74194 105 221 98 4042 75 9194 408 728 4549 74195 95 298 185 4044 88 4408 728 4554 74196 95 298 185 4044 88 4408 728 4558 74197 85 Complete 4046 128 4411 728 4554 74198 150 28 4006 15 4058 4415 785 4558 74LS* CMOS* 4050 48 4415 785 4558 74LS 4000 15 4050 72 4435 825 4559 10 20 4005 105 4052 73 4435 825 4559 10 20 4005 105 4054 119 4450 255 4588 13 33 4008 87 4057 2576 4452 255 4558 11 33 4008 87 4057 2576 4452 255 4558 11 4 75 4009 50 4059 480 44990 525 4581	П								99	4541	
74194 105 241 238 4043 84 4409 220 4553 74196 95 298 185 4044 88 4409 720 45554 74196 99 386 86 4045 145 411 120 4554 74198 150 15 range available 4047 97 44157 785 4558 4415 785 4559 4559 4559 4559 4559 4559 4559 45	3			221	98	4042				4543	
74195 85 298 185 4044 88 4409 720 4553 74197 85 Complete 4046 128 4410 720 4556 74197 85 Complete L5 range 4047 77 4412V 1380 4556 74195 150 4000 15 4051 72 4422 545 4558 4550 4560 4560 4560 4560 4560 4560 4560	П	74194		241	236	4043				4549	
74198 99 386 86 4045 148 4410 720 4554 74198 150 Complete 4046 122 4411 1049 4557 74198 150 Cavailable 4046 38 4415 78 4557 74LS* 74LS	: 1	74195	95	298	185				720		
74197 85 Complete 1044 525 4411 1040 4556 1050 1050 1050 1050 1050 1050 1050 1	1	74196	99	386	86			4410			
7418* 150 LS range available 4047 87 4412V 1340 4557 4415F 785 4558 4559 4455 4558 4559 4558 4559 4558 4559 4558 4559 4558 4559 4559	1	74197	85		te			4411			
Available	8		150					4412V	1380		
74LS*	П	11100		availabl				4415F	785		
74LS* CMOS* 4050 48 4419 284 4560 4561 4560 4561 4562 4562 4562 4562 4562 4562 4562 4562	Н			availabi	•					4558	
74LS0" CMOS* 4190 48 4422 2845 4560 01 18 4000 17 4052 72 4433 1099 4561 02 20 4002 17 4053 72 4440 1275 4562 01 04 22 4005 105 105 105 128 4450 295 4562 10 20 4007 18 4057 2570 4452 295 4568 10 20 4007 18 4057 2570 4452 295 4569 12 475 2570 4452 295 4569 12 27 28 4010 50 4060 115 4490V 325 4581	4				_					4559	
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27 28 4010 50 4060 115 4490V 525 4581	5	13	38	4008							
27 28 4010 50 4060 115 4490V 525 4581		14	75	4009	50	4059					
					50	4060	115	4490V			
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SWITCHES* WATFORD ELECTRONICS 299 TOGGLE 2A 250V SPST 28 DPST 34 DPDT 38 4 pole on off 54 nt Displays 675 TiL211 Grn TiL212 Yellow C·2" Red O·2" Yellow, Grn, Amber ORP61 (Continued from opposite side) DIODES OPTO ISOLATORS TIL111/2 AAZ15 BA100 BY100 BY126 BY127 TIL114 TIL117 OR - 1.5. TORS* Plastic (TO220) case | 1A DPDT | 13 | -ve 0 · 5A · 5V · 6·2V · 8 | -ve 0 · 5A · 5V · 6·2V · 9 | -ve 1A 5V · 12V · 95 | -ve 1A 5V · 12V · 12V · 95 | -ve 1A 5V · 12V · 12V · 95 | -ve 1A 5V · 12V 160 148* 75 12 VOLTAGE REGULATORS* TO3 Can Type 1A +ve: 5V, 12V, 15V, 18V 145p each LM309K LM309K LM323K 598 MVR5 or 12 150 LM320LM320LM320LM320LM320LM320LM320LM320LM320LM320LM320LM320LM320LM320LM320LM320-CRO33 OA9 OA47 OA70 OA79 OA81 12 12 15 OA85 1A -ve: 5V, 12V 226 95 105 Plastic Case: +ve 285 0-1A (TO92) 5V, 6·2, 95 8·2V, 12V, 15V 30 1A (TO220) 5V, 12V, 15V, 18V, 24V 85 0A91 0A95 OA200 OA202 IN914 ZENERS Rng: 2V7-39V 400mW 9p IN916 IN4001/2* VEROBOARD* Pitch 0-1 0-15 ROCKER: (white) 5A 250V SP change-over centre off 38 VEROBOARD* 0-1 (copper 2½ × 3½" 41p 2½" × 5" 43p 3½ × 5" 43p 3½ × 5" 58p 2½ × 17" 195p 4½ × 17" 195p Pkt of 36 pins Spot face cutter Pin insertion tool 0-15 clad) 33p 45p 45p 60p 121p 163p 0.15 (plain) 28p 22p — 28p Rng: 3V3-33V 1·3W 17p IN4003* ROCKER: (Illuminated, red) Chrome Bezel 5A 250V SP 52 IN4004/5* IN4006/7* ROTARY: "Make-A-Switch" Make your own multiway Switch. Adjustable Stop Shafting Assembly, Accommodates up to 8 Wafers VARICAPS MVAM115 120 BA102 25 BB104 40 BB105B 40 BB106 40 IN4148 IS44 38p 78p 187p 165p 30p 85p 98p 45p 18 20 27 3A/100V* 3A/400V* 3A/600V* Mains Switch DPST to fit Break Before Make Wafers, 1 pole/ 12 way, 2p/6 way, 3p/4 way, 4p/3 way, 5p/2 way 3A/1000V* 6A/600V Noise Diode Z5J 160 VERO WIRING PEN* + Spool 325p Spare Wire (Spool) 88p; Combs 7p ea. 325p Spacer and Screen SCR's* ROTARY: (Adjustable Stop) 1 pole/2 to 12 way, 2p/2 to 6 way, 3 pole/ 2 to 4 way, 4 pole/2 to 3 way 41 FERRIC CHLORIDE* 11b bag Anhydrous 65p + 30p p. & p. ALUM. BOXES 1A50V DALO ETCH RESIST PEN-Plus spare tip 1A100V 1A200V ROTARY: Mains 250V AC, 4 Amp

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5AQ5	0.75	6AU6	0.50	6CS7	0.85	6U4GT	0.80	35A3	0.70	ECH83	0.60	EY8i	0.50	PCF82	0 - 45	UBF80	0.60
5AT8	0 · 80	6AV6	0.75	6CU5	1.00	6V6GT	0.65	35B5	0 · 65	ECH200	0.80	EY87	0.50	PCF84	0.45	UBF89	0.60
5T4	0.75	6AW8A	0.75	6CU6	1.00	6X4	0.60	35C5	0.70	ECL80	0.60	EY88	0.55	PCF86	0.75	UBL21	0.85
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5X4G	0.80	6BF5	0 · 85	6DO6B	1 45	I2AD6	0.80	*90C1	1 - 20	ECL85	0.65	G Z30	0.65	PCL84	0.75	UCH8I	0.65
5X8	0.90	6BF6	0.75	6DT5	i · 00	12AE6	0.85	*807	1.00	ECL86	0.85	GZ32	0.65	PCL86	0.85	UCLBI	0.70
5Y3GT	0 65	6BG6G	0.30	6DT6	0.80	I2AF6	0.80	*811A	3.80	EF80	0.40	GZ33	3 80			UCL82	0.75
5Z4GT	0 · 65	6BH6	0.85	6DT8	0.80	12AJ6	0.70	*829B	8.80	EF85	0.48		4.50	PCL805	0.75		0.75
6AB4	0.55	6B16	1.20	6DW4	0.90	I2AL5	0.65	*832A	8.20	EF86	0.60	KT66		PD510	3 · 35	UCL83	1.00
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6AF4A	0-80	6BQ7A	0.65	6GK5	0.70	12AU7	0.47	DF96	0.60	EFI84	0.70	OB3	0.75	PL84	0.75	UM80	0.60
6AG5	0-65	6BR8A	1.20	6GK6	0.90	12AV6	0.85	DK92	1.00	EFL200	1.20	OC2	I · 40	PL95	0.70	UM8I	0.75
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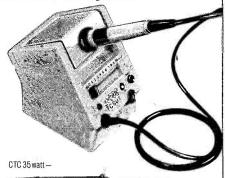




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