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 Latching relay maline operitad \＆e\％o contact： 2．11．Malna oposated threo 10 Atmp changeovers open tyot one zeraw fixing ci－2．Mnsy othar
typeat with diferent coil yotans and sontact arrengemente are In atock，enqutrles livilod．


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| All thame bave 20.54 w | Hz Pimary |  |  |
| :---: | :---: | :---: | :---: |
| VOLTAG厚 | CURRENT |  | PRteg |
| ${ }_{2}^{14} 4$ | \％Amp | TM | S1．40 |
| 4 | 7 amp | TN32 | 2．310 |
| Or | \％amp | TM3 | 4 |
| d．5v | 3 mmp | TH\％ 37 | 全 |
| 8．5Y | 200 ma | TM 91 | 11． |
| 8－5y－0－65y | 10 mma | TM 81 | 81. |
| －5v－0－6． $5 v$ | 780 mA | TM 7 | 4． 1 |
| 8－3v－a－8．3y | 100 ma | TH 33 | cite |
| $5 \cdot 34$ | 2 mmp | TM 4 | ［1． 11 |
| 8．5y | 1 amp | TH2 | Ef． 6 |
| sivy＋8－5v nep winding | \％ | TM 12 | 61．42 |
|  | 1 mmp | 1M ${ }^{\text {¢ }}$ | E1．${ }^{4}$ |
| \％ | I amp＇c＇cot＊ | TM6 | 4t |
| Or | stamp | TM11 | 27 |
| gr | 8日mp | TH ${ }^{38}$ | Cis 24 |
| 10\％ | 23.10 | TM 15 | C4． |
| 40v－0－50v | $12 \%$ amp | TH is | 84 |
| 12v－0－12y | 4 mmp | T127 | 動． 12 |
| \＄2v | 1 amp | TM ${ }^{\text {P }}$ | ct．${ }^{\text {c }}$ |
| $13 \%$ | \％amp | TM | 4． 4 |
| 184 | 1 mmp | TM 10 | 64．6 |
| 12v－0－12y | 50 ma | TM | Etic |
| 1400－12v | 1 amp | TM 41 | 28.24 |
| 15v tappod 9V | 2 amp | TM 11 | c2．74 |
|  | 7 mmp | TM堂 | 隹號 |
| 1． $\mathrm{v}-0-15 \mathrm{y}$ | stiamg | TM 27 | ctiz |
| 15v－0－18v | 3 Samp | TM 3 | 4.8 |
| 17 v | tomp | TH 12 |  |
| 17y | \％mo | TH15 | 515 |
| 90\％ | $t \mathrm{cmp}$ | TM14 | 51．82 |
| 20v | 5 mmp | TM 27 | ${ }^{4} 8$ |
| 20 V | 521 mmp | TM 15 | 如 |
| 年v－0－20y | 8 amp | TM 16 | E4 ${ }^{4}$ |
| $13 \%$ | 100 ma | TM 29 | ＜1， |
| 244 | 11 mmp | TH 16 | ［8． 12 |
| 248 | 9 amp | T117 |  |
| $24 y+5 r 7$ mmp | 2 smp | TM 38 | C2 31 |
| 24 v | 4 \％rpp | T\＃40 | C4， |
| 25 | 1tamp | TM | 24 4 |
|  | 2 tmp | TM 3 | C2， |
| 300 tapped 24، 20． 15412 | 31 mmp | TM 27 | E4． |
| $30 \%$ | 8 amp | TH15 | c4．tat |
| $37 \%$ | 37 nmp | TM 3 | 5 Et |
| 40v tapped at 30r，20\％ 410 c | \％amp | TM 15 | 24t |
| $5 \mathrm{Sow}-2$ amd with B＇3v ahroud |  | TM 92 | Es－＊ |
| S0y | 8 Emp | TM ${ }^{\circ}$ | 61145 |
| 60\％ | 5 mm | TH 24 | ct 0 |
| 76v－s amp wilin $8 \cdot 3 \mathrm{~Sv}$ Ehroud | ed | TM ${ }^{\text {明 }}$ | 钴 16 |
| 75 y ， | if amp | TM 24 | 57.2 |
|  | 4 smp |  | 77.18 |
| 100 y | 1 amp | TM 26 | c） 7 |
| 100N－0－100w | $t \mathrm{mp}$ | TH | 67.6 |
| 130\％tepped 120\％ | famp | TM88 | ¢37 |
|  | ＋ Fmp | TH ${ }^{25}$ | 478 |
| 950y－0－250\％with e－3y 2A | 50 mA | TM 38 | 578 |
| 250\％ | 100 mA | T0438 | d 73 |
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| AAY30 | － 13 | ASY2］ | \％ 5 | 8 C 67 | 0 17 |
| AAY32 | －13 | ASZis | $1 \cdot 23$ | BC． 170 | $0 \cdot 4{ }^{\circ}$ |
| AACi3 | － 2 | ASZ1a | 1.2 | $\square{ }^{\text {B }}+5$ | $0 \cdot 1{ }^{\prime \prime}$ |
| AZ15 | － 31 | AS217 | 125 | $\mathrm{BCl}^{172}$ | －19＊ |
| AZ1 | － 25 | AS220 | － 75 | BCis | 615 |
| AC407 | 075 | ASZ21 | 15 | 8 Cl 7 | 48 |
| ${ }_{\text {ACl }} 125$ | － 36 | Alsic | 1－79＊ | 8 C 978 | 013 |
| AC528 | 025 | AU113 | $170{ }^{\circ}$ | $8 \mathrm{Cl79}$ | 0 碞 |
| AC127 | － 25 | AปY10 | 9．76＊ | EC5 ${ }^{\text {c }}$ | － $11^{\circ}$ |
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| ACs6 | － 20 | GA14： | $0 \cdot 110$ | BC $\$ 8$ | － $122^{\circ}$ |
| AC14iK | － 35 | 8A154 |  | 8 C 2 | $0.14{ }^{\prime \prime}$ |
| AC42 | 0.20 | BA15 | － 62 | BC21 | d $14{ }^{*}$ |
| Ct42 | － 30 | BA156 | $0 \cdot 13$ | 8 C 21 | ＋97＊ |
| A 478 | － 23 | BAW62 | $0 \cdot 6$ | BC837 | － $17{ }^{*}$ |
| ACtar | － 25 | BAX13 | $0 \cdot 1$ | $8 \mathrm{Cza3}$ | $0 \cdot 12^{\prime \prime}$ |
| C\％789 | － 23 | EAX16 | ． 67 | 8 C 301 | 048 |
| Acys | － 0 － | C107 | 12 | 3C303 | 0 （t） |
| ${ }_{\text {A }}{ }^{\text {cris }}$ | $0 \cdot 65$ | 8C109 | 42 | 8C30 | －20 |
| Y19 | － 5 | Bcros | $0 \cdot 13$ | BC3 | －tc＊ |
| CYa | $6 \cdot 4$ | BC：13 | －130 | 8 C 327 | $0 \cdot 22^{\circ}$ |
| ACY21 | ＊ 65 | EC314 | ¢ 1 1＊ | 日Ca28 | 0．11＊ |
| ACYO | f． 55 | BC115 | － 5 ¢ | BC，337 | 0 11＂ |
| AD849 | ¢ 76 | BCtI5 | 418 | 日c3s8 | 0.410 |
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| C9L31 | 156 | ECCBa | 650 | EL42 | f $75 *$ |
| cess | \％＊＊ | \＆ccast | －55＊ | E¢81 |  |
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| DaF81t | ＊ $40 *$ | ECCO1t | © $55^{*}$ | Etst $\dagger$ | 0 58＂ |
| DAF96 | 1．6＂ | Ecctas | 4．00＇ | Elst | 2－Es＊ |
| DFitt | 40＊ | ECFB0］ | －60＂ | ELSSt | － $0^{6}$ |
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| DKEt | －${ }^{\text {E }}$ | ECH35 | $200^{\circ}$ | EMEO | 1．14＊ |
| OKP2 | 1．25 | ECH42 | 43＂ | EM64 | 1－98＊ |
| OKP | 1．10＊ | ECHE1 | － $56{ }^{\circ}$ | EMS4 | 1 時 |
| DL92 | －．75＊ | \＆CHB3 | －${ }^{\text {ct }}$ | EMA5 | ＋25＂ |
| DL94 | $480{ }^{18}$ | ECHRO | 6．65＂ | EMa7 | 1．59＊＊ |
| DL98 | $810{ }^{4}$ | EClset | － $68{ }^{\circ}$ | ENP94 | 4． 55 |
| DYgot $\dagger$ | $84{ }^{*}$ | ECLİ | － 35 | EY54\％ | － $78{ }^{4}$ |
| DY802 | －19＊ | ECL63 | $158{ }^{\circ}$ | EYgBt | $00^{\circ}$ |
| EBSCCt | 100 | ECLBO | 0．65＊ | E240 | 1．25＊ |
| EABCa |  | ECLL80 | 7．${ }^{\circ}{ }^{\circ}$ | E241 | 124 |
| EACil | 8．56 | EF37A $\dagger$ | ${ }^{3} 60^{\circ}$ | EZ801 | － |
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| Ebc33 | 1．75＊ | EF50 | －${ }^{\text {ct }}$ | G234 $\dagger$ | 152 |
| EBC4 | 125 | EF80 | 6．45＂ | KT81 | 3－40＊ |
| EBCA | $11{ }^{\circ}{ }^{\circ}$ | Efs3 | \％．74＊ | KT\％ | 180 |
| ERCtG | －65 | EFES | $0{ }^{58}$ | KT88 | 4 3 |
| E8F80 | － $45^{\circ}$ | EFF\％ | ［45＊ | KTWS1 | 175 |
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| E8Fag | $4.40^{\circ}$ | EFIT | － 55 | KTW6 | $175^{\circ}$ |
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| ECCSIt | － 59 | EF183 $\dagger$ | － 50 | OA2 $\dagger$ | 45 |
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| OY\％${ }^{\text {a }}$ | 52 p | EF09 | 5 sp | ${ }^{P}$ | 真p | PCFEOS | Ex． | PL81a | ${ }^{4} \mathrm{p}$ | U日F6s | p |
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 pait threa yoara the mimplifier has betn rofnad to the extent thet it must bi one of the most tolabit and robuat High Fidelify modulas in ins Warid．
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## High quality audio modules for Stereo and mono






| GUTPGT POWER | T Watts RMS |
| :---: | :---: |
| LOAD IMPEDANCE | 8 ohms． |
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| FAEQUENCY RESPONSE | 50 Hz to $20 \mathrm{hHiz} \pm 3 \mathrm{cEs}$ |
| TONECONTROL RANGE | $\pm \pm 2 \mathrm{dBs}$ at 100 Hz and 10 hHz |
| SENSITIVITY | 100 mVior tull output |
| INPMT EMPEDANCE |  |
| TRANS FORMER REOUIREMENTS | 20．V．A．C．rated at 1A |
| DIMENSIONS | $200 \mathrm{~mm} \times 130 \mathrm{~mm} \times 33 \mathrm{~mm}$ |

Tha Sterac 30 comprises a complete stereo pre－amplifas，power ampllifer，thid power aupply．This，with only the eddition

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| AL60 <br> AUDIO <br> AMPLIFIER MODULE <br> 25 Watt 8 MS <br>  <br> $+\mathbf{1 2} \%$ VAT | OUTPUT POWER <br> SUP＂ply <br> LOAD TMPEDANCE <br> TOTAL HARMOONTC DISTCRTION <br> EREOUENCY RESPONSE <br>  <br> MAX．HEĀ̄ SUNK TEMPERATÜRE EIMENSTONS | 25 Watte RM5 <br> $30-50 \bar{V}$ <br> $\bar{B} \overline{\mathrm{E}} \mathrm{C}$ ohm： <br>  <br>  <br> $2 B 0 \mathrm{mV}$ for full output <br> $00^{\circ} \mathrm{C}$ <br>  |
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| This high qually audio anplifler modulatis for use to 25 RHS with diatortion levets below 0．1\％． | a aquipment and atarec amplifit | d provides astput powers un |



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35 Watte RMS
SUPPLY $40-60 \mathrm{~V}$
$8-18 \mathrm{hm}$
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20 Hz to 30 kta （Typicaidy 260 mV for tull autput $260^{\circ} \mathrm{C}$
$90^{\circ} \mathrm{C}$

The Also ia almilar In design to the AL60 above and ta of the same hioh quatity bul provides outpul pawers up to 35W with distortion leveft helow $0.1 \%$


This unti，dasignated AL250，is a powef amplifier proulding an oulgut of up to 125 W RMS，Into a 4 ofm load．

| A3ids 10y | MAXIMUM SUPPLY VOLTAGE POWER OUTデラ゙ラ for $2 \%$ THD | 30 V 10 Wntte RMS |
| :---: | :---: | :---: |
| Auplo R．zh．s． | TOTAL MARMONIC DISTORTION | Leas than $25 \%$ |
| AMPLIFIER | LOAD IMPEDANCE | $\frac{6-18 \text { ohm }}{100 \mathrm{Kohms}}$ |
| MODULES | INPUT IMPEOANCE <br> FRËQUENCY REXPONSE | roo K ahme <br>  |
|  | 9ENSITIVITY | 75 mV lor full gutput |
|  | प्1लENSIONS | $74 \mathrm{~mm} \times$ ¢ 0 mm $\times 8 \mathrm{mmm}$ |


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## SPMBO

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| PNPUT A，C，VOLTAOt | S5－40V |
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| 万ा心 | $108 \mathrm{~mm} \times \mathrm{Bimm}$ 30mm |

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## PA100

3TERED
PRE－AMPLIFIER

## £15．80

$+12 \mathrm{P} \% \mathrm{FAT}$
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| FREQUENCY AESPO | 20 Hz to $90 \mathrm{hHz} \times 1 \mathrm{~dB}$ <br>  |  |
| :---: | :---: | :---: |
| TOTAL HARMONR DISTO |  |  |
| SENSITIVITY inputs <br> TAPE <br> g．RADIO TUNER | $\$ 0 \mathrm{mV} 100 \mathrm{~K}$ ohma $100 \mathrm{~m} / \mathrm{il} 10 \mathrm{~K} \mathrm{~K}$ ahme 3．mV／80 K ohme | $\begin{aligned} & \text { Forsm } \\ & \text { output } \end{aligned}$ P50 |
| TQuillisation |  |  |
| BASS CONTROL RANGE |  |  |
| TREBLE COMTROL RANGEE | ＋ $10-20 \mathrm{~d}_{\text {de }}$ at 15 kHz |  |
| SIGNALINOISE RATIO | Botte：Ihin 65 dBE ［All Ingutio） |  |
| INPUT OVERLOAD | Eettier than 26 dBe（All inputi） |  |
| SUPPLY | 201040 V |  |
| DIMENSICNS | $300 \times 80 \times 33 \mathrm{~mm}$（lese | controlos |

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## MPA30 <br> magnetic cartridge

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magneltc enrtridge with your
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amplifier enabling magnetlc carifloges io to ueg whera lackitiot ekfat for the uss of caramic cartrideon only．

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## PA12

STEREO

## PRE－AMPLIFIER

## The PAl2 Sterec Pre－

Amplliser chassis in desıgned snct recommended for une with in AL 20 P30 Audlo Amplifer Modulas recommended for une with the T638 Trangormes fatures toclude onfof wosumer lupply and the and Treble controls Complete with lspe outaul
FFEQUENCY RESPONSE $20 \mathrm{~Hz}-20 \mathrm{kHz}$ t -3 d 日） GASS CONTROL TREBLE CONTROL TNPUT HMPEDANCE INPUT SENSITIVITY $\quad-\frac{1 \mathrm{Meg}+\mathrm{ohm}}{300 \mathrm{mV}}$ CROBSTALK SIGNALINOISE RATIO OVERLOAD FACTOR TAPE Ö̈TPUTIMPEDANCE DTMENSIONS
3.5 mV for 100 mV otrput

Within $\pm 1$ die from 20 Fixt to 20 kHz
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18 so $30 \mathrm{~V}-\mathrm{e}$－ rlth


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## BACK NUMBERS

We are very glad to announce the re-establlshment of a PW Back Numbers Service for our readers. In future back numbers dated from June 1977 onity will be avaifable from our Post Sales Department for 65 p , which Includes postage and packling. Cheques and Postal Orders should be made payable to IPC Magazines Ltd.
Send your orders to:- Post Sales Department, IPC Magazines Ltd., Lavington House, Lavington Street, London SE1 OPF.

IN a recent lecture entitled "Tomorrow's Broadcasting-The Technical Possibilities", Dr. Borls Townsend, head of the Engineering Information Service of the Independent Broadcasting Authority, highlighted some examples of the pitfalls in trying to forecast future developments in engineering. In 1924 for example, Campbell Swinton, talking to the RSGB, dismissed the topic of television (or as he called it then, seeing at a distance) as ". . . probably scarcely worth anybody's whlle to pursue". In more recent times, a well-known television engineer swore an affidavit to the FCC in America, that the shadow-mask colour display tube could not be mass-produced!

In making his forecasts, Dr. Townsend saw the biggest advances in broadcasting coming from the ever-widening use of sophisticated microcircuits. Their use in signal-processing circultry would allow good qually TV pictures to be produced from scenes with low lighting levels, and reproduced from small, cheap videotape recorders. The adoption of microprocessor-based control systems has already allowed the IBA to operate and maintain 400 transmitters with the same number of staff as were needed for only 40 transmitters some dozen years ago. The introduction of similar systems into studlos is also having far-reaching effects.
Sticking my own neck out, I foresee that the domestlc TV recelver itself is likely to undergo a change of use over the coming years. The growing popularity of TV games, such as the Tank Battle which we feature in this issue, is the first step in this process. The broadcast Teletext services, Oracle from the IBA and Ceefax from the BBC, are already established and hopefully we will soon see a reduction in the price of receivers fitfed with the necessary decoders. Also on the data front, the Post Office has recently announced that their Viewdata service is to be made available to the public from January 1979. All this will mean that the TV set will be used less as a source of broadcast entertainment and more as a source of information and participative entertainment. The adoption of microcircuits should also allow the domestic videotape recorder to be reduced in mechanical complexity and hence in cost, so that we may be able to view our favourite TV programmes when it suits us, rather than when the planners deem that we should.

Geoffrey C. Arnold

## PLEASE NOTE-CORRESPONDENCE

We do not operate a Technical Query Service except on matters concerning constructional articles published In PW. We do not supply service sheets or Information on commercial radios, TV's or electronic equipment. All querles must be accompanfed by a stamped self-addressed envelope otherwise a reply cannot be guaranteed.

## Bally date

'The Northern Mobile Rally 1978', organised by the Otley Radio \& Electronics Society (G8JTD, G3XNO) is to be held at The Victoria Park Hall, Keighley on Sunday 21 May between 11,30 and 17.30 .
Talk in stations on 2 m f.m. S22 and 70 cm f.m. SUB. There will be trade stands, films for the children, bar, refreshments and many other attractions. Further details from:
J. E. Annakin, G8DFZ, Rally Manager, 25 Ashfield Place, Otley, W. Yorks LS21 3dN.

## Revival '78

The Martlesham and 1pswich Radio Club and The Ipswich Area Civil Service Sports Association (I.A.C.S.S.A.) are again organising an outdoor event for the Radio Amateur and his family, to be held on Sunday, 14 May 1978, at the I,A.C.S.S.A. Sports Ground, Straight Road, Bucklesham (NGR TM 222 421). Special attractions include; A v.h f./u.h.f./s.h.f. aerial gain competition and demonstration, measurement of transmitter and recelver performance, bring and buy stands, Ad-hoc trading tables at $£ 1$ per hour for visitors and home-brew equipment competition.
The South Anglian Repeater Group, West Suffolk f.m. Group and Raynet
will be well represented and a 2 m (R3 S22, 70 cm ) (RB4 SU8) and h.f. bands talk-in station will be in operation, probably using the call sign GB3 SWR. Further attractions include; Big name traders, demonstrations of Viewdata iv and microprocessor games, vintage wireless displays, raffle, pistol and archery ranges, flying display of radio controlled aircraf:, plus many other family entertainments.

The event will start at 11.00 am , admission 40p (accompanied children free). A licensed bar will be open from midday and there will be snacks, teas etc., throughout the day.

Further information can be obtained by sending a SAE to: C.P. Ransom G8LBS, 79 Camden Road, Ipswlch, Suffolk IP3 8JN.

## Tirst shew

The Dept. of Electrical Engineering Science at the University of Essex is organising the 1st Essex Electronics Exhibition on 18/19 April 1978.
Admission is by free ticket issued by either the companies exhibiting or by the Department. Further details from:
E. P. Strudwick, Dept. Electrical Engineering Science, University of Essex, Wivenhoe Park, Colchester, Essex. Tel: 020644144 Ext. 2248.

## Hi-Fi Seminar

Latest developments in turntables and pickups, amplifiers, loudspeakers, tuners and tape recorders will be reviewed in a one day seminar being organlsed by the Society of Electronic and Radia Technicians. Speakers include such famous names from the audio field as James Moir, James Linsley Hood, John Borwick, Angus McKenzie and Basil Lane. The lectures will be followed by demonstrations.

The seminar will be held at the Institute of Marine Engineers, Mark Lane, London EC3 on Wednesday, 7 June 1978, commencing at 10.00 am . Fees are £15 ( $£ 10$ to SERT members) and this includes a copy of the papers, coffee, lunch and afternoon tea.

Further details from: SERT, 8-10 Charing Cross Road, London WC2H $0 H P$.

## Can I help you!

Are you the secretary, arganiser or general dog's body of your local radio club or any other group whose functions may interest readers of PW. If so, let me know and I will endeavour to publicise your rally, get-fogether, whatever, through this column. Remember though, we compile the magazine some time ahead of publicacation day (e.g. this note was witten in mid-March), so, the earlier I can have details, the better.

Alan Martin

## HIDIL IDIE

Radio 2 Tuner, July 1977
On $p 213, \mathrm{C} 7$ is incorrectly shown as $3 \cdot 3 \mathrm{pF}$. This should be $3 \cdot 3$ aF as shown in the components list.

## "Shoot", August 1977

Certain errors in the circuit diagram on p. 283 have been noted. The PCB is correct, however. VR4'should be 22 k preset, and not $2 \cdot 2 \mathrm{k}$. Pin 14 of IC3A should be connected to +8 V line as should pin 14 of IC4A and pin 3 of IC2B. Pins $1,2,5,6$ and 7 of IC3C should be connected to OV line, as should pin 7 of IC4B. Under the heading "Connection" on p.282, it is claimed that the sync. output is taken from pin 16 of ICl of "Tele-Games" unit. This should read "pin 15".

## IC of the Month, Sprague ULN-3006T Hall Effect

 Switch, March 1978 PW the second paragraphfollowing the heading "The Hall Effect", p. 845, and commencing "The current carricrs..." is somewhat ambiguous, and should read "The current carriers in the silicon (which may be electrons or "holes") are both deflected to one side of the material, depending upon direction of current flow and magnetic field, in accordance with Fleming's Left Hapd Rule."

## "Multi-Range Test' Meters," Mareh 1978

Page 839, the paragraph commencing "The minimum measurable. . " should read:

The minimum total circuit resistance necessary, if full-scale deflection is not to be exceeded, is therefore 1500 ohms, This will be made up of the meter movement resistance plus a series current-limiting resistor, both of which are internal to the instrument. External readings are from zero upwards.

Experimenter's Corner, p. 910 April 1978. In the circuit diagram and text for "LED Light Display", the pnp transistors are incorrectly shown as ACl76 (npn). These should of course be AC128 in every case.


## Construction

In general the construction is not critical and the prototype built by the author used plain Vero matrix board as shown in the photographs. However a complete set of p.c.bs is available and the various drawings show these and the associated component placements.

Leads from the front panel controls to the boards should be of screened wire, and the millivoltmeter circuit board should have a tinplate screen fitted around it as detailed in Fig. 8. It can be made from tinplate cut from a cocoa tin and is held in position by two paper clips soldered to it as shown. The better the screening of boards and components, the lower the final distortion measurement limit will be.

## Initial Setting Up of the Meter

Set all pre-sets to halfway. After carefully checking that no mistakes have been made in the construction, switch on. At first the meter will swing about for a few moments and then settle down.

Allow about one minute before setting up the distortion meter as follows:-
Disconnect the bridge output lead to the millivolt meter attenuator and then connect an audio generator directly to the millivolt meter attenuator. (The point originally connected to the bridge.)

Set the output of the generator at 1 kHz to give full scale deflection of the meter with the millivolt meter range switch on the IV range, Switch to the 10 V range, meter should fall to $10 \%$ of full scale. If not, adjust the pre-set 3 for more feedback and repeat test. Only the minimum amount of feedback required to obtain a linear scale should be used. When the reading drops to $10 \%$ of full scale, pre-set 2 should be adjusted so that a 5 V input gives full scale on the 10 V range, and 0.5 V input gives full scale on the 1 V range, etc.
Switch millivolt meter attenuator back to 10 V range and transfer generator to the input socket. Reconnect output lead of bridge. Set/Read switch to Set. Adjust generator output for 10 V and distortion meter input attenuator to maximum. Adjust VR7 in bridge circuit for full scale on meter. (If your
generator does not have 10 V output, use 1 V and switch millivolt meter attenuator to IV range.)
Turn input attenuator to minimum. The residual reading on meter should be less than 0.4 mV on the lmV range with all screens etc., in place. The authors instrument has a residual noise 0.24 mV on the lmV range. This represents a 93 dB measuring range for signals above 10 V , i.e., down to $0 \cdot 003 \%$. Set generator to IV output at 1 kHz . Set distortion meter to IV range and adjust input attenuator for a convenient reading about two thirds of full scale. Switch generator to 1 V output at 100 kHz . Adjust trimmer TCl for the same reading on meter as before. This adjusts the frequency response of the meter for a flat response to 100 kHz .
Switch generator back to 1 kHz and set its output to 10 V (or 1 V if the higher output is not available).
After adjusting the input attenuator for full scale deflection, switch Read/Set switch to Read. Adjust bridge frequency and balance controls to obtain lowest possible reading on meter, reducing the voltage range switch as the meter readings reduce. The final lowest reading obtained is the Total Harmonic content of the Test Signal.

## Operation

A typical set up is shown in Fig. 13. When measuring very low values of distortion, it is very important to avoid multiple earth connections. Only the amplifier under test should be earthed. The other test equipment must have its normal mains earth removed and connected only via its connecting lead to the earth of the amplifier. Care should be taken that all test equipment is safe and fitted with mains isolation transformers or battery operated. Multiple earths can cause very high distortion readings and could be very misleading.
Before commencing a measurement, connect your generator directly to the distortion meter and measure its distortion. The figure you get from this test will set the minimum distortion you can measure. This minimum should be restricted to at least twice the direct reading obtained if reasonable accuracy is to be maintained.
To make a measurement, adjust the input signal of the amplifier under test to provide the required


Inside view of the completed instrument showing the layout of the p.c.bs and the controls mounted on the front panel. The power supply is fitted to the cabinet base and is the only board not mounted onto the front panel
output into a dummy load. Set the distortion meter millivolt switch to a suitable voltage range. Adjust the input attenuator for full scale reading ( $100 \%$ ). Switch to "Read" and null out the fundamental signal. As the optimum bridge balance is obtained, the millivolt meter range will have to be switched to the next lower range. Read off the distortion direct from the meter when no further reduction in level is possible. Note that the frequency and balance controls are interdependent.
With an oscilloscope connected to the socket provided on the distortion meter the harmonic content of the signal can be examined. A scope sensitivity of approximately $10 \mathrm{mV} / \mathrm{cm}$ is required. With inputs
below 10 V the minimum distortion readable will be reduced.
(Note: For inputs of less than 10V set the millivolt meter voltage switch to the range which will allow full scale deffection to be obtained. This range then becomes [for purpose of measurement] the $100 \%$ range and all other ranges move down by the same factor. For example, with IV input, switch meter to 1V range, this is then $100 \%$ distortion full scale, the $0 \cdot 1 \mathrm{~V}$ range becomes $10 \%$ full scale etc.)
For accurate measurements the limits are shown in Table 1. However measurements of distortion to lower levels can be made, but with decreasing accuracy-
Nearly all pre-amplifier and tape recorders have

This view shows the components and boards mounted onto the back of the front panel. This is the prototype unit using matrix boards instead of p.c.bs



Fig. 5 : Component placement drawing for the Bridge Circult p.c.b. (Board 1)


Fig. 7: Copper track layout for Board 3. Fig. 8: (Below) Component placement drawing for Eoard 3



Fig. 6 : Copper track layout for Board 1. Ready drilied boards for this instrument are available from Reader's PCB Services (See page 68)


Fig. B: Details of the tinplate screen for Board 2. The small bracket is soldered to the copper earth tracks at the top of Board 2


Fig. 10: Copper track layout for Board 2
Fig. 9: Component placement drawing for the Meter Circuit p.c.b. (Board 2). Note the small tinplate bracket soldered to the earth track


Component placement drawing for the Power Supply p.c.b. (Board 4)
outputs in excess of 1 V and power amplifiers of the hi-fi type will have more than 10 V output. So the t.h.d. meter should enable the hi-f amplifiers to be checked down to below $0.01 \%$ with reasonable
accuracy. Do not be surprised if your amplifier does not reach the lowest figures at the extremes of the audio band and do not maintain the highest frequency for longer than it takes to make a measurement.


Fig. 11 : Copper track Iayout for Board 4. A ready drilled set of boards for this instrument Is available from Reader's PCB Service (see page 88)

Fig. 12: Detalls of the components mounted difectly onto $\mathbf{s}_{2}$ waler tags, The drawing hows the components openedzout fop clarty, they should be arranged along the ax ls of that owftch mectandsimin cylindrical fashion


Fig. 13: A typical set up for testing an audio amplifier using the Audio Distortion Meter


The following procedure is to assist the operator when using the distortion measuring meter for the first time. It is confined to measuring the distortion of a 1 kHz signal, and should help the operator to become familiar with the basic operation.

| Minimum Input | Accurate Meaisurement |
| :---: | :---: |
| 10 V | $0.00 \%$ |
| 1 V | $0.6 \%$ |
| 100 mV | $1 \%$ |
| 10 mV | $10 \%$ |
| 1 mV | - |

## Table 1.

The controls should be set as follows:Meter Range Switch on IV range.
Frequency Range Switch on range 3.
Filter
Set/Read switch on Set
Input Attenuator at Zero
All other controls Midway
Connect a sine wave signal of 1 kHz , IV RMS to the input socket. Advance the input attenuator until meter reads $1 V$, i.e., full scale. Switch the set/read switch to the read position and rotate the frequency dial for minimum reading on meter. Adjust balance control to reduce meter reading further. Switch the volt meter range switch to next lower range as the signal is reduced by adjustment of the frequency and balance controls.

When no further improvement (reductions) can be obtained the distortion can be read directly from the

Resistors

| $\pm$ W $5 \%$ metal oxide |  |  |
| :---: | :---: | :---: |
| $33 \Omega$ | 1 | R43 |
| 1008 | 3 | R7, 49, 52 |
| $560 \Omega$ | 1 | R6 |
| $1 \mathrm{k} \Omega$ | 1 | R50 |
| $1 \cdot 2 \mathrm{k}$ \% | 1 | R24 |
| 2-2kS | 3 | P5, 8, 23 |
| 2.7ks | 1 | R9 |
| 4.7kS | 2 | R10, 45 |
| 10h\% | 9 | R14, 15, $\ddagger 6,18,20,21,36,46,48$ |
| 12k5 | 3 | FR3, 42, 51 |
| 22kS | 2 | R40, 53 |
| $27 \mathrm{k} \Omega$ | 1 | R41 |
| $33 \mathrm{k} \Omega$ | 5 | R26, 27, 28, 29, 30 |
| 39 k ת | 1 | f2 |
| 47kS | 3 | R1, 19, 24 |
| f00ks | 5 | Rt1, \%2, 13, 25.47 |
| 150kS | 1 | R38 |
| 180kS | 1 | R17 |
| $220 \mathrm{k} \Omega$ | 1 | R22 |
| 270k』 | 1 | R39 |
| $3304 \Omega$ | 1 | R44 |
| 470kS | 1 | R37 |

1 W 2\% metal oxide

| $1 \Omega$ | 1 | R35 |
| :--- | :--- | :--- |
| $10 \Omega$ | 1 | R34 |
| $100 \Omega$ | 1 | R33 |
| $1 \mathrm{k} \Omega$ | 1 | R32 |
| 10 ks 3 | 1 | R31 |

## Potentiometers

$\frac{1}{4}$ inch diameter spindles

| $100 \Omega$ | 1 | VR9 |
| :--- | :--- | :--- |
| $1 k \Omega$ in. | 2 | VR2, 6 |
| $10 k \Omega \operatorname{lin}$. | 1 | VR8 |
| $10 k \Omega+10 k \Omega$ iin. | 1 | VR4,5 |
| $20 k \Omega$ in. | 1 | VR3 |
| $100 \mathrm{k} \Omega \log$. | 1 | VR1 |

Miniature horiz. skeleton preset
$100 \mathrm{k} \Omega$
1 VR7

Semiconductora
Diodes
OA202 $\quad 4 \quad$ D1, 2, 3, 4
IN4001 $4 \quad D 5,6,7,8$
BZYB8C24
D9

Transistors
BC433B 9
TR1, 2, 3, 4, 5, 6, 7, 8, 9

## Capacitors

Polyester

| 2.2 nF | 2 | $\mathrm{C} 7,12$. |
| :--- | :--- | :--- |
| $0.01 \mu \mathrm{~F}$ | 2 | $\mathrm{C} 6,11$ |
| $0.047 \mu \mathrm{~F}$ | 2 | $\mathrm{C} 5,10$ |
| $0.1 \mu \mathrm{~F}$ | 1 | C 25 |
| $0.22 \mu \mathrm{~F}$ | 8 | $\mathrm{C} 4,9,13,20,21,22,23,24$ |
| $0.68 \mu \mathrm{~F}$ | 1 | C 15 |
| $1 \mu \mathrm{~F}$ | 2 | $\mathrm{C} 3,8$ |

Electrolytic Printed circuit board mounting

| $2 \cdot 2 \mu \mathrm{~F}$ | 63 V | 3 | $\mathrm{C} 14,18,27$ |
| :--- | :--- | :--- | :--- |
| $10 \mu \mathrm{~F}$ | 63 V | 6 | $\mathrm{C} 2,16,17,28,29,31$ |
| $100 \mu \mathrm{~F}$ | 63 V | 1 | C 30 |
| $220 \mu \mathrm{~F}$ | 63 V | 1 | C |
| $470 \mu \mathrm{~F}$ | 63 V | 1 | C 33 |

Electrolytic Axial teads

| $470 \mu \mathrm{~F}$ | 63 V | 2 | C 19.32 |
| :--- | :--- | :--- | :--- |
| $1000 \mu \mathrm{~F}$ | 63 V | 2. | C 26.34 |
| $4700 \mu \mathrm{~F}$ | 25 V | 1 | C 35 |

Ceramic trimmer
$3-35 \mathrm{pF} \quad 1 \quad \mathrm{TCl}$

## Switches

| Min. foggle s.p.d.t. | 1 | S3 |
| :--- | :--- | :--- |
| 2p. 6w. midget wafer | 2 | $\mathrm{~S} 1,4$ |
| 2p. 6 w. minlature rotary switch | 1 | S 2 |
| Mains switch assy to fit S mech. | 1 | S 5 |

## Miscellaneous

Transformer 24V 20VA Miniature
Case RS 509 - 888
Printed circuit boards (Four in set) Readers PCB Service. 24 V Indicator lamp
Knobs Sifam collet fixing type 15 mm diameter
W151 wing knob (3)
Kt50 plain knob (3)
K151 plain knot with line pointer (2)
N150 nut covers (6)
C150 caps (8)
Figure dial for 15 mm knobs with pointer ( 3 )
numbered 1-10(1)
Meter 1 mA f.s.d. $90 \times 74 \mathrm{~m}$ m approx.
BNC $50 \Omega$ sockets (2)
Tinplate sheet for screen (cut from used cocoa tin or similar)
PW Front panel overlay (Obtainable from PW Editorial Office)
meter and voltage switch.
That is, if initial full scale ( $100 \%$ ) was IV and final reading was (say) 6 mV , then distortion is $0.6 \%$. The low frequency filter can be switched in for measuring frequencies above 1 kHz if hum is affecting the measurement.

## Other uses of the Meter

With the input attenuator at maximum, the meter can be used as a normal $A C$ millivoltmeter with full scale range of 10 and $1 \mathrm{~V}, 100,10$ and 1 mV . This would make it possible for example to measure the
output of a magnetic pick-up directly.
It can also be used for measuring Hum and Noise in an amplifier. By adjusting the input attenuator for full scale on a test signal from an amplifier and then removing the test signal and shorting the test amplifier input to earth. The meter will then indicate the residue Hum and Noise of the test amplifier, for example: if full scale was obtained on 10 V range with the test signal and then after removing it the reading was (say) 6 mV , this represents a ratio of 1666:1, approximately 64 dB .
Many other uses can be found and a few hours spent using the meter will be very rewarding. A REVIEW OF RECENT DEVELOPMENTS in general, the author does not have any more infornialion on prociucts thaty apmeats in the article

## I spy Strangers

Some kind soul sent me a whole heap of papers from the recent International Solid State Circuits Conference and there seems to be some real goodies on the way (not available yet). One which took my eye is a single 14-pin dual-in-line package which houses a complete motion detector. It is intended for application in electronic toys. This little beasty can be made to keep an eye on a 2 ft , diameter area at 8 ft . The photodiode itself is actually integrated as part of the chip. An external loudspeaker is required and when connected up, the unit will emit a whooping noise whenever it senses motion within its "sight" area. It carries on making this din for a set period of time, then it goes back to sleep and wails for the next "something" to move within its sighting area. It would seem to offer great possibilities as a burglar alarm, etc.

A further nice feature of this little i.c. is that it has another mode of operation. To change it to this you need only add a single connection.
In its new mode it will "search". It does this by flashing an external bulb at some 3 Hz . At the same time, it croaks out a random series of squeaks and grunts (the paper offers the more sophisticated description, "emits audible notes"). When the "thing" detects its own reflection it immediately sounds an alarm and simultaneously increases the bulb flashing rate to $25-30 \mathrm{~Hz}$. The chip uses two technologies; linear bipolar, and $1^{2} \mathrm{~L}$.

## Hi-digi-fi

A recent report from Japan details feverish activity among audio equipment manufacturers-in the digital field. It is now virtually certain that the hi-fi systems of the future will be digital.
To date it seems that the only "standard" to emerge is a wide acceptance of a 30 cm disc as the norm. Interestingly, though, the early professional systems will use tape before moving over to disc, and it is expected that the consumer scene will also follow this pattern.
But don'f get too enthusiastic about digital audio. The world concensus of opinion is that it will be some four years before professional digital audio
systems really catch on, and a further six years after that the consumer market will blossom. It could therefore be some ten years before you see these systems advertised. The main hold up will be price. Initially, systems will be expenslve and the first few years will be needed to gradually bring the prices down.

Why go digital anyway? It seems that analogue hi-fi has now gone about as far as it car, whereas digital is in its infancy. In terms of improvements digital technology has much to offer the audio buffs. To start with, frequency response and dynamic range are both independent of the characteristics of the tapes or discs used. it is also claimed that there is no crosstalk problem between channels. From all the specifications on systems that l've read, the responses are flat (very, very flat) right up to 20 kHz , and dynamic range, even at this early stage is well over 80 dB -some 20 dB better that most analogue systems I read about these days.

Another advantage of the digital approach is wow and flutter-there isn't any! This is because all the signals are retimed so accurately ©uring playback.

Various individual technologies are to be employed initially, including a laser/disc system. But the comforting thing is that despite the very different techniques, the final product, be it tape or disc is compatible in that one can convert material from tape to disc and vice versa.
Perhaps we'll all end up with a home computer to play our gramophone records on. Wonder what that dog, squinting down that trumpet/horn thinks about it all?

## Useful Chips

Another chip which could be very useful for the home constructor (when it becomes available) is a new level detector i.c. Onto the chip the manufacturers have managed (somehow) to cram five comparators. a voltage regulator, five output driver fransistors, five scaling resistors and an input buffer stage with a high input impedance. By connecting five l.e.d.s (plus usual limiting resistors), each l.e.d. can be made to light as the input voltage increases in steps of 200 mV , i.e. for each 200 mV input,
the next l.e.d. illuminates. The opencollector outputs on the chip can handle currents up ta 80 mA and voltages up to 18 V . In practical terms this means versatility because the ratings allow not only l.e.d.s to be used as indicators, but also filament lamps. By using suitable circuitry, the device can be made to flash the first lamp or I.e.d. continuously when the input level falls below the 200 mV threshold level.

## Charge!

Charge those c.c.d. (charge-coupled devices) are in the news again-well worth keeping an eye on. This time it's a Japanese company that is using c.c.d.s in an experimental colour television camera. Each of the three c.c.d. chips (one for each primary colour has an array of 111,192 separate little sensing elements in an area $10.3 \mathrm{~mm} \times 9.1 \mathrm{~mm}$. If small is beau iful, then these devices must be fantastic.
The colour television camera, when it comes on the market (probably late next year if all goes well) will come complete with zoom lens, built-in camera control circuitry, and electronic viewfinder, Price is set at around the $£ 500-£ 600$ matk. Weight will be less than 2 kg .

## Programmer

Microprocessors are here to stay and many are available to home constructors. One of the problems is learning how to programme these clever little electronic beasties. An answer is offered by a German manufacturer. He's marketing a little "black box" which can be used in conjunction with a black and white TV receiver. The black box has a light pen and the TV receiver is used as both input and output terminal. The box comes with a 230 -page manual and costs around £280. For a further $£ 140$ (approximately) the purchaser can add a cassette control Interface for writing (and reading) memory data on standard, commercially avallable tape. The reading/writing rate is some 1 kilobyte in 90 seconds.


# © 'purbeck' 



## Part 3

## IAN HICKMAN



Having made up the Stabilisers board, check it over and mount it in the mainframe. Set all preset pots. to mid travel. The Raw Supplies have no current limit protection of course, so unless you feel $100 \%$ sure that everything is going to be all right, the following procedure is suggested.

Either run up the voltage between pins 9 and 5 to +17 V using a current limited lab. bench power supply or if one of these is not available, connect up the 0 V line between Boards I and 2, but connect the +17 V Raw supply from C19 to Board 2 pin 9 via a $330 \Omega$ resistor. Check that +5 V appears at pin 8 and that the voltage at pin 10 responds to varying VR202.
If so, all is well, though it may not be possible to set the output at pin 10 to +12 V until the $330 \Omega$ resistor is removed. The +17 V raw may now be wired in permanently and VR202 set to give +12 V at pin 10 . Similarly, check out the -6 V and -12 V stabilised outputs at pins 3 and 2 with the -17 V raw from C20 supplied to pin 1 via a $330 \Omega$ resistor, then wiring up and setting VR203 for -12 V at pin 2. Finally, connect the +300 V raw supply from Board 1 to pin 6 of Board 2 and check that +150 V appears at pin 7. Note that due to the absence of load current drawn from the +150 V stabilised supply the " +300 V Raw" will be nearer +360 V .

This completes the checking out of the instruments supplies and at this stage it is worth completing the mainframe and front panel wiring. Wiring confined to the front panel should be completed before offering it up to the mainframe and likewise, wiring of S1 and S2 should be completed before offering them up
to the front panel. Note that the probe power socket SKT10 is also used as a distribution point for the stabilised supplies to various controls on the front panel.

A ten way colour coded ribbon cable brings the supplies from the rear of the instrument, five of the leads terminating on SKT10 as detailed in Fig. 2 Front Panel Wiring. Further Iengths of ribbon cable, stripped down to the same five coloured leads distribute the stabilised supplies from SKT10 to the edge connector of Board 4 and thence to that of Board 3 . (Note that the pins of both edge connectors are numbered 1 to 36 working from the bottom upwards. As both Boards have the components facing outwards when plugged in, the edge contacts on the component face of the Board read from right to left for Board 3 and left to right for Board 4.) The remaining five wires in the 10 -way ribbon cable are used as follows: Black. 0 V from Board 1 pin 6 to common earth point at Board 3 edge connector.
Brown. Pins 1 and 2 of c.r.t. (cathode) to clockwise end of VR5 track.
Violet. Pin 5 of c.r.t. (anode 1) to wiper of VR6 (focus).
Grey. R18 and D1 (mounted on c.r.t. base) to wiper of VR5 (brilliance).
White. -800 V Stabilised from Board 1 pin 7 to anticlockwise end VR5 track.
The earthed pins on the edge connectors of Boards 3 and 4 are each individually wired direct to a 16 s.w.g. tinned copper wire running the length of the edge connector, at each end of which it is supported on a


## KMOO4

Fig. 1: Power supply interwiring diagram and c.r.t. base connections. This drawing should be read in conjunction with Fig. 2 Front Panel wiring diagram. Please note that RV5 and RV6 should be read as VR5 and VR6
solder tag under the edge connector mounting screws.
Thus Board 4 ground plane, when plugged in, is earthed to the base of the mainframe via the edge connector mounting bracket and to the front panel via the 4BA pillar. Board 3 , whilst similarly earthed, picks up the Black wire from the ribbon cable, at a point on the $16 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. earthwire adjacent to the Input Low contact and from the same point an earthwire runs to the earth point on S3 and thence to a solder tag under SKT 1 mounting screw.
This earthing arrangement is essential to avoid instability, ensure a flat $Y$ amplifier frequency response and avoid ringing on square waves.
When making the connections between the front parel and mainframe, lay the panel down in front of the mainframe as though it were hinged at its lower edge. (This is why Fig. 2 Front Panel Wiring has been drawn the way it has.) Dress the wires from

## WARNING <br> Extra care must be taken when working on any part of this instrument while power is switched on, 1100 volts can kill. When delving into the insides of the scope for any reason with power on keep one hand in your pocket

the front panel along towards the "hinge" and thence off to their destinations. This will ensure that when the front panel is offered up and secured in position, there is adequate but not excessive lead length.
Having rechecked the wiring and removed the temporary link from Board 1 pin 6 to chassis, plug in briefly and check straight away that all the stabilised voltages are correct, indicating no shorts anywhere. Check that the slider of VR5 covers the range -750 V to -800 V approximately and that of VR6 -350 V to -600 V approximately. The c.r.t. base wiring can now be completed, except for C18 and the deflection plates.
In fact, the c.r.t. and mu-metal screen can now be fitted and a simple check carried out if desired. To do this, temporarily connect one end of a $47 \mathrm{k} \Omega$ resistor to +150 V stabilised and the other end via a $100 \mathrm{k} \Omega$ resistor to chassis.


Fig. 2: Front panel back wiring. This diagram has been drawn with the top edge of the fronf panel at the bottom so as to correspond with its orientation when placed on the bench for the purpose of wiring. Connecting wires from the front panel to other parts of the instrument should be long enough to allow the panel to be lowered to the bench in front of the mainframe. This makes for easier working conditions. Please note that potentiometers labelled RV1 to $\mathbf{5}$ should be read as VR1 to 5 and also that the $\mathbf{2 2 0 k} \Omega$ Focus potentiometer at the bottom righthand corner of the diagram labelled RV5 should be VR6. This drawing should be used with Fig. 1 the power supplies interwiring diagram
Resistors
（U＇nless otherwise specified $5 \% \%$ carbon film）

|  | R13 10k |
| :---: | :---: |
| R2 inoker | R14 5－6kS |
| R3 10M $10 \%$ | R15 680 |
| R4 $1 \mathrm{MS} \mathrm{1} \mathrm{\%}$ | R16 330kı2 |
| R5 $910 \mathrm{k} \Omega 1 \%$ | R17 j00k |
| Roi 1MA $1 \%$ | R18 foms $10 \%$ |
| 27 TM＠ $1 \%$ | R19 688 |
| R8．1 M $1 \%$ | R20 100 S |
|  | f2\％ 1008 |
| Rio $10 \mathrm{k} 51 \%$ | R22 1008 |
| R11 ths \％\％ | R23 1002 |
|  | R24 1－2k ${ }^{\text {d }}$ |

## Capacitors

C1 $0: 1 \mu \mathrm{~F} 350 \mathrm{~V}$
C13： $0-1 \mu$ F $1 \% 63 \mathrm{~V}$
C2 lopF Ceramic
C3 2－22pF
C4 2－22pF
C5 2－22pF
C6 2－22pF
C7 2－22pF
C8 $2-22 \beta F$
C9 $2 \cdot 22 \mathrm{pF}$
CiO．2－22pF
Cil $0 \cdot 14 \mathrm{~F} 350 \mathrm{~V}$
C12 $1 \mu \mathrm{~F} \% 6 \mathrm{~V}$
C14 10nF $9 \%$ 63V
C15 $1 \mathrm{nF} 1 \% 63 \mathrm{~V}$
Ci6 5－65pF．
C17．4．7 $\mu \mathrm{F}$ 100V
C18 0．1 2 F 1000 V
0192500 LF 25 V
$\mathrm{C} 20.2500 \mathrm{p} \boldsymbol{\mathrm { F }} 25 \mathrm{~V}$
C21 47pF Cerainic
C22 470pF Ceramic
C23 4：7nF met．film C24 47 mF met．film
Potentiometers
$20 \%$ Mas $\frac{2}{2} W$ 交inch shafls
$\checkmark \mathrm{F} 125 \mathrm{~K} \boldsymbol{2}$
VR4． $1 \mathrm{k} \Omega$
VR2 $2 \cdot 2 \mathrm{k} \Omega$
VR5 47k
VR3 10kå
VRG $220 \mathrm{k} \Omega$
Hindúturs
${ }^{1} 1$ See Text
1．2 See Text
Diodes
Di． 1 N 4148

D2 Hi brightness lie．d．
Tansiormer
 7554 （Barrie Electronics）
Miscellaneous ：
Fi ldunch x dinchi A fuse and holder
Si Miniature s．picio．toggle switch
S2 Miniature push button n．o．
S3 2p 5w 2 wafer（Maka－switch）
\＄4 2p 6w 1 water．
＂\＄5 Miniature s．pnc．o．toggle switch
S6， 7 Miniaiture s．p．c．o．toggle switch
SKT1 BNC socket，round s0』（U心t094／U）
SKT2 4 mm sócket，black
SKT3 4 mm souket，biue
SKT4 4 m m socket，white
SKT5， 0.4 mm socket，green
SKT7 4 mm socket，yellow
SKTB， 9 4mm sucket，red
SKT10 5 pin DAN $180^{\circ}$（A）
CRT 3EP1 plus base
Case mouriting clips（verl）for C 19,20
Eage cannector 0 it inch pitch 36 way 2 off
Knubs Stiam 15 mm ，colfet fixing with nut covers and caps．
$\begin{array}{ll}\text { K150 p登in } & \text { 20f7 } \\ \text { K151 line pointer } & 4 \text { off } \\ \text { W151 wing and Ifne pointer } & 6 \text { off }\end{array}$


Connect all four deflection plates to the junction of the two resistors，i．e．approximately +100 V ．With VR5，VR6 and VR201 all set to midtravel，switch on and allow a few seconds for the tube heater to warm up．

Adjusting VR5 should produce a spot on the end of the tube and VR6 should enable it to be focused to a small diameter．If at either side of this setting it looks elliptical，wider or taller as VR6 is adjusted， VR201（astigmatism）should enable this to be cor－ rected．Mind where you put your hand when adjusting VR201，it＇s not far from the e．h．t．on Board I！It should be possible to focus the spot down to a pin－ point，provided the brilliance control is not advanced too far，though of course VR201 will need resetting when Boards 3 and 4 are fitted．
In fact Board 3，the $Y$ amplifier is the next step and full details of this will be published in next month＇s instalment．

Several readers have enquired about the possi－ bilities of using alternative tubes for Purbeck． We cannot advise anyone as to the suitability of components other than those specified．Not only will the mechanical construction need alteration，but revised amplifiers and e．h．t． súpplies will also be required．


## Introducing DIGITAL LOGIC

Beginning a primer in logic, which will cover gates, flip-flops and counters. characteristics of TTL and CMOS and design techniques.


A simple circuit design which provides an accurate 200 kHz output signal, phaselocked to the BBC's Droitwich transmitter, useful for calibration of receivers and digital frequency meters.


When one thinks of people in another part of the world, one usually imagines that their life style is almost identical to one's own. The fallacy of this concept is only obvious when one travels and is exposed to another way of life. Contrary to popular belief, radio amateurs are no different to other people and although the hobby is international, it takes on different forms in different countries. This article is an introduction to amateur radio in the United States of America.

## The Transmitting Licence

It is much easier to become a radio amateur in the USA than in the UK. There are several classes of Licence, each having different examination standards and frequency privileges, as shown in the table. In general, Novices have to pass an examination in elementary theory and a 5 w.p.m. Morse test. The examination is administered by a radio amateur volunteer on behalf of the licensing authority (which in the USA, is the Federal Communications Commission).
The Novice licence is valid for two years and until recently was distinguished by the WN prefix. It allows its owner c.w. only privileges in the $10,15,40$ and 80 metre bands with a maximum power input of up to 250 watts.
The Technician class of licence requires slightly more technical knowledge than a Novice, and allows operating privileges similar to our Class B licence on segments of the 6 metre band and higher frequencies. The General, Advanced and Extra class licences are all allowed all-band, all-mode operation, but each class of licence (except Extra) is limited to segments of the band. The test for the General requires 13 w.p.m. c.w. and the test for the Extra requires 20 w.p.m. Increasing levels of technical knowledge are required for up-grading from one class to the next.

The licences are free and except for the Novice, are valid for five years. Separate mobile or TV licences are not required. The segments of the bands available to the different classes of licence are summarised in Fig. 1.
The band split between phone and c.w. is decided by the FCC for the amateurs, and not by the amateurs themselves as in most other parts of the world. In general the input power limits are 1 kW input d.c. or 2 kW input p.e.p. except on Top Band, which is segmented in both frequency and power depending on geographical location as shown in Fig. 2.

## The HF Bands

Operation on the h.f. bands is very different to that in Europe. The vast majority of the stations appear to be using the full legal limit and beam antennas. Thus the bands are crowded with strong signals all originating from the states, and it is difficult to hear non-stateside signals at times. The USA is so large that in general, any foreign station is DX.
The bands are so crowded that if you want to work the states from outside, you should get up into the General parts of the bands when they are open to the USA. You may then be giving the stations you work, their first G contact. If they want your QSL they will probably offer to QSL direct and even send you IRCs. There are many more of them than there are Gs, so if they are not in a rare state such as Utah or Delaware, let them QSL first. If they want your card, they will. If you operate in the Extra or Advanced segments of the bands the chances are greater that you will be working someone for his umpteenth G contact.

The $3 \cdot 5 \mathrm{MHz}$ band is so wide $(3 \cdot 5-4 \cdot 0 \mathrm{MHz})$ that the c.w. part is called 80 metres, but the phone part is called 75 metres. At this point of the solar cycle, it

TABLE: US Amatour Radio Classes

| cłasas | Morse <br> Requirements | Techinical Knowledge Required | Dperating, Prififeges |
| :---: | :---: | :---: | :---: |
| Novice | 5 w.p.m. | Hardly any | $\begin{aligned} & 10,45,40 \\ & \text { and } 80 \mathrm{~m} \text {. } \end{aligned}$ |
| Technfician Genere! | $\begin{aligned} & 5 \text { w.p.rm. } \\ & \text { \$3 w,p,m. } \end{aligned}$ | About RAE level About RAE leyel | 6 m and higher All bands, some frequencles |
| Advanced | 13 w.p.m. | A 班tlé more than RAE fevel | All bands, more frequencles |
| Extra | 20 w.p.m. | As Advanced | All bands, all frequencies |

offers cross continent contacts late at night. The 40 metre band has similar characteristics but is little used at night due to the vast amount of broadcast station interference. Sectors of the band allocated to broadcast stations in Europe are allocated to amateurs in the Americas and those broadcast stations come in loud and clear in the USA. The 20,15 and 10 metre bands are pretty much the same as in Europe, in terms of distance worked, but are generally without the language barrier, because the common language in the states is English (more or less). There is thus very very little incentive for the American amateur to learn a foreign language.

## Traffic Handling

American radio amateurs have third-party traffic handling privileges. This means that they can pass messages for people other than radio amateurs. For example, a station in New York can contact a station in Los Angeles and ask that station to pass on a message to a non-amateur. He can even have him connect his radio to the telephone line via a phone patch and make a radiotelephone call, thus saving on his long distance telephone bill. There is a number of countries that have reciprocal agreements with the USA about third-party traffic; in other words they allow traffic to be passed between their country and the states. Thus on all bands one can hear a number of nets passing traffic messages. Once a year the American Radio Relay League (ARRL) organises a traffic handling contest (called Sweepstakes), in which the information exchanged simulates message traffic.

## The VHF Bands

At v.h.f. there is no 4 metre band, but there are operating privileges at 50 MHz ( 6 metres) and 220 MHz ( $1_{4}$ metres). Six metres opens up to sporadic $E$ much more often than 4 metres, and thus a lot of the activity is on c.w. and s.s.b. It has
properties very similar to 10 metres in terms of ground-wave communications capabilities, but DX is of course much more scarce on 6 than on 10 .

The 2 metre band is 4 MHz wide ( 144.148 MHz ). The top two megahertz are filled with f.m. repeaters and simplex channels spaced 30 kHz apart. Small segments of localised s.s.b./c.w. activity exist close to 144 and 145 MHz , but in the main the lower two megahertz comprise the wide open spaces. At the time of writing, the FCC is proposing to open some of it up for repeaters. There is OSCAR-related activity at about 145.9 MHz . Thus apart from narrow and sparse areas of activity at 144,145 and $145 \cdot 9 \mathrm{MHz}$ the lower two megahertz is an uncharted wilderness at this time. Local s.s.b. or c.w. activity on a nationwide basis is rare.

In most big cities, tuming the low end of the band by day will be very unrewarding with little to be heard. Even in the evenings you would be lucky to hear more than three simultaneous contacts taking place in the low half of the band, but during a contest a tremendous number of stations suddenly crawl out of somewhere and fill up the one or two hundred kilohertz. However, in the major cities the f.m. channels will be crowded. In most big cities all the repeater channels between 146 and 147 MHz are in use as well as some of the 147.148 MHz ones. There is no f.m. calling channel as such, just find a repeater and use it seems to be the rule.

In the states, the pioneers on v.h.f. set up repeaters to extend the range of their converted taxicab f.m. mobile equipment. As newcomers came on the band, they found the repeaters in existence and joined in. If people did not like a particular machine they were free to build and use their own on an adjacent frequency. In some parts of the country there were even repeater "wars" over choice frequencies between two repeater groups, each trying to force the other group to change frequency (this was before the days of synthesisers, when everyone was crystal controlled). Eventually voluntary frequency control was established by area-wide organisations. In the main how-


Fig. 1: HF Frequency Assignments in the USA. Note that higher class licensees have privileges in lower class segments, the table shows the lowest class allowed in each segment


Fig. 2: Top Band frequency allocations in the USA
ever there have been few cases of deliberate interference with repeaters, because the newcomers to v.h.f., especially to 2 m , generally only used f.m. and soon learned the advantages of the extended range and continuous monitoring of the repeaters.
In the UK the situation is different, the bands were in use before the advent of the repeaters. Also the 2 m band is only two megahertz wide and everyone has to fit into it. Hopefully it will take just a short while for everyone to find out the advantages of repeaters and common sense will then prevail.

The 220 MHz band is similar in characteristics to 2 m . There is little in the way of s.s.b./c.w. commercial equipment for the band, and hence most of the activity is $\mathrm{f} . \mathrm{m}$. The same applies to 70 cm . Here, most of the activity is f.m. between 440 and 450 MHz , i.e. right at the top of the band, so that conversion of surplus equipment involves the minimum of changes. The 70 cm band is also used for remote control of lower band repeaters and for inter-repeater links. Activity on higher frequencies is at par with Europe, namely very rare and due to only a few individuals.

## Using Repeaters

Many f.m. repeaters are connected to telephone lines. This allows for "auto patch" facilities, whereby amateurs can actually access the telephone network via the repeater and dial calls using the tones. They can report accidents to the police, call home and ask
the wife if they should stop off at the local.supermarket and pick up some groceries, or do as one radio amateur did over one of the local machines here in Washington DC; while sitting in the garden by the side of the pool, he used his walkie talkie equipped with a touch-tone pad to dial his house phone and ask his wife to bring him out another can of beer!
The number of repeaters is constantly growing. The ARRL publishes an annual directory that is free for the asking to members. The frequencies are based on a 600 kHz split with a spacing of 30 kHz between channels. They are known by the kilohertz values, i.e. a repeater on $146 \cdot 25 \mathrm{MHz}$ (in), $146 \cdot 85 \mathrm{MHz}$ (out) is commonly known as the $25 / 85$ machine. In the $146 / 147 \mathrm{MHz}$ region the input frequency is the lower one, whilst in the $147 / 148 \mathrm{MHz}$ region the reverse is true and the higher frequency is the input channel. This was carefully arranged this way so that receivers could be peaked up at 147 MHz and work with the whole range of channels. These frequencies are of course not allocated to the amateur radio service in Europe, and on my last trip to the UK, I found that some of the American repeater output channels that I had in my rig were in use by the police.

Apart from f.m. the majority of activity on 2 m and 70 cm seems to involve OSCAR. Project Oscar started the whole thing in California with the launch of the OSCAR I satellite in 1961, and AMSAT took over in 1969. The ARRL puts out a lot of free

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educational stuff about the OSCAR satellite programme, which by-the-way is available to anyone worldwide, for the asking. Worked All States via OSCAR is just possible for Eastern stations and is well within the capabilities of anyone in the continental United States west of the Mississippi river.

## Equipment

Salaries in the states are generally at between two and four times the level of the equivalent salary for the same job in the UK. Thus on a basis of hours worked, equipment is much cheaper over here. The large number of amateurs support a few manufacturers so that there is quite a variety of domestic gear available as well as the ubiquitous Japanese black boxes.

Parts for the homebrewer are also readily avail. able by "mail". Many suppliers advertise in the various magazines, and their wares can be ordered by post, or by telephone quoting credit card numbers.

Crystals for 2 m transceivers are available over the counter in most major cities and the prospective visitor can wait until he arrives in the USA to purchase the bulk of his crystals. If he is lucky the store may even have facilities for tuning the rig frequency. If you are interested in operating in the USA when visiting this country you can obtain detailed information from the RSGB, or write to the FCC, Washington DC, 20554 for an application form (Form 610A). Make sure that you send in your application at least three months before your trip, because it will take them that long to reply.

## Listening and Viewing

The receiving side of the hobby is also somewhat different. The medium waves, f.m. and v.h.f. TV bands are filled in most major urban areas. Medium wave DX-ing is a little easier than in Europe, because most of the stations broadcast in English. Stations are spaced 10 kHz apart, which enables the intercontinental DX to creep in between the cracks. The majority of stations are east of the Mississippi. The FCC recognise the sky-wave effect at night and regulate the band occupancy such that many local stations are required to close down at local sunset. This allows clear-channel stations to be heard over continental distances at night. It is thus for example possible to drive through downtown Los Angeles in the evening, tune the car radio a little and listen to broadcasts from Denver, Oklahoma and Iowa, or to drive around in Washington DC and listen to stations in Montreal, Chicago and New Orleans.
The v.h.f. TV bands are also full in most cities with some additional channels in the u.h.f. bands in use. All transmissions in the Americas are 525 -line 30 -fields, thus there are no modifications necessary to the TV sets for their use in TV DX-ing. The f.m. band is 20 MHz wide in the range $88-108 \mathrm{MHz}$, with 88.92 MHz being reserved for public broadcasting stations. These stations, usually run by universities or local authorities, broadcast educational programmes, classical music and selected shows from the BBC ' (including the Goon Show) as opposed to the popular music or news churned out by the commercial stations. There are three major networks in the states that broadcast commercial programmes, and each city usually has a local outlet of each as well as one or more independent stations and the public station. This allocation applies to television as well as radio. Recent broadcasts on public televsion have included, I Claudius and Upstairs-Downstairs.

Many of the independent TV stations broadcast old films and TV shows including UFO, The Saint, The Prisoner, The Avengers and Danger Man. The f.m. stations usually broadcast in stereo. Both medium wave and f.m. stations usually churn out popular music, each specialising in a particular type, and there may be one or two stations that continually broadcast news. The u.h.f. TV band is relatively sparsely populated, its growth being curtailed by cable television systems.

## Citizen's Band

Closely related to amateur radio is Citizen's Band Radio. The Citizen's Band is a small allocation of spectrum space at about 27 MHz . It seems to have been originally allocated for personal communications between a fixed base station and a mobile (so that for example, a husband could talk to his wife on the way home from work) or between mobiles, so that the drivers of two cars travelling together can communicate.

Licences are now available free for the asking, with no tests involved. Power input is limited to 5 watts d.c. a.m, and equipment is cheap. A fortychannel transceiver can be purchased for about $\$ 50$. There are estimated to be millions of CB stations in service, the majority of them being unlicensed.

Homo sapiens is a creature of invention. The American branch of that species, perhaps more inventive than other branches, has devised new uses for the Citizen's Band. Lorry drivers use it to warn each other of impending police radar speed traps. Prostitutes have been known to solicit customers via CB radio. Hobby operation including QSL'ing abounds and most stations use self-given "handles" rather than their official call signs, always assuming that they have an original call sign.
nllegal power amplifiers are often used and some operators are even equipped with amateur band equipment such as Yaesu FTIO1s. In the cities the channels are overcrowded and communication ranges are limited by the numbers of conversations taking place on the frequency. One channel has been set aside for emergency use (ch 9) and one is used as a calling channel (ch 19). Almost everyone monitors channel 19 when on the road unless they have QSYd for a particular reason to another channel, and contribute traffic and police location information. Hitchhikers also solicit rides over channel 19. One neighbour of mine estimated that in two weeks of use he recouped the cost of the equipment due to the timely warning he received about the locations of police radar speed traps that would otherwise have caught him. Citizen's Band is a boon if used correctly, but if abused is a mess.

## Microprocessors

The latest arrival on the electronics hobby scene is the home computer based on the microprocessor. Thousands have been sold and clubs are forming all over the country. Microcomputers can be the subject of many articles in themselves, but suffice it to say for now that their use is invading the home as well as the amateur radio shack. It is estimated that within five years the vast majority of homes in the US will contain at least one microprocessor.
This article has been an introduction to amateur radio and the electronics scene in the United States. Future articles will go into more detail about the various aspects of the hobby.

The General Instrument AY-3-8710 integrated circuit is a' 625 line interlaced "TV Tank Battle Game" for two players. The "battleground" consists of white barriers and "a series of black mines. There is one white and one black tank, each controlled by two single pole double throw paddle switches biased to centre off. A push button for each player controls gun firing and a push button allows the battle to be reset. Pin 22 of the i.c. is switched to control the tank traps. In the open position the tanks can drive through these barriers, in the other position (grounded) the tanks halt when they collide with them. Motor sounds are provided for each tank as well as gun fire and shell explosions, and the score is coded to each tank. The tanks are driven like real tracked vehicles, pushing both switches forward causes the tank to go forward. If the switches are held the tank automatically speeds up after a few seconds. If the switches are released the tank continues at the speed reached at the time of release. Pulling both switches back causes the tank to reverse, while holding one switch back and one forward causes the tank to rotate. To stop the tank when it is going forward momentary selection of reverse is required. The shell has a range of about two thirds of the screen and after firing there is a reload period before you can fire again. The shell can be steered during its flight by rotating the tank. The shell will pass over the mines but will explode on hitting barriers. A hit on your opponent counts one point, while running over a mine counts one against you, When one player reaches 16 hits the scores flash to show that the game has ended.

## Circuit Description

The circuit diagram is shown in Fig. 3. Tl provides 12-0-12 volts which is full wave rectified by D1 and D2 and smoothed by Cl. IC1 regulates the supply VP to approximately 6.5 volts, VR1 adjusting the voltage and C2, 3 and 4 providing decoupling. IC4 a and b provide the 4 MHz clock to the AY-3-8710, L1 being adjustable to allow the oscillator to be set to the correct frequency. IC3 $a$ and $b$, with their
associated Rs and Cs, provide the shaping for the fire and explosion sounds. IC3c does the same for the motor sounds which ane all fed via IC3d to the output transistors TRI and TR2. Switches S4, 5 and 6 control the left tank and S7, 8 and 9 control the right tank. S3 is the game reset and S2 is the tank trap switch. R2 to R4 mix the video signals and the composite video is buffered by the emitter follower TR3 and fed to the modulator.

TR4 and its associated components act as the modulator which runs at a frequency of approximately $160-170 \mathrm{MHz}$, with harmonics extending into the u.h.f. band.

## Construction

Construction is relatively straightforward, most of the components being mounted on the p.c.h. It is, however, advisable to use sockets for IC2, 3 and 4. The component layout is shown in Fig. 7. Drill 6BA clearance holes for board fixing, FSI, T1 and IC1, and a 6 mm hole for L1. Drill other holes to suit the component leads. Before mounting any components on the board place the p.c.b. in the box with a modulator at the front right, and the p.c.b. about 10 mm from the right side of the case. Drill 4 fixing holes in the bottom of the case using the p.c.b. as a template. Also drill a 6 mm hole in the bottom to line up . with Ll core. Put the p.c.b. to one side and drill the box and front panel as shown in Fig. 1. Also drill a few 6 mm holes in the base and rear of the case for ventilation. Stick a small piece of speaker cloth over the rear of the speaker hole on the front panel, fix the speaker in place with epoxy adhesive and mount the switches S1, 2 and 3 . The sound output is controlled by VR3 which is mounted on the front panel.

Assemble the p.c.b. using the layout Fig. 7 and parts list as a guide but do not fit IC2, 3 or 4 into their sockets yet. R9 can be either a $20 \mathrm{M} \Omega$ resistor or two 10 MO resistors (R9a and R9b). Fit a T05 heatsink to TR2, cut and bend a piece of aluminium sheet $60 \mathrm{~mm} \times 35 \mathrm{~mm}$ as a heatsink for TR1 as shown in Flg. 4, drilling through the p.c.b., and retaining 1 Cl and the heatsink with a 6BA screw. Cut and bend

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## BATTLE GAME


a piece of timplate (cocoa tin) as shown in Fig. 5 to form a box for the modulator screening, fit the sides and bottom by soldering to four Veropins as shown in the drawing but leave the top plate off until the


## A

Fig. 1: Front panel drilling diagram

Fig. 2: Main printed circuit board connections $>$
unit is working and displaying a picture on the television screen.

Carefully check the p.c.b. for correct assembly and freedom from inadvertent shorts such as solder




## AFig. 4: Heat sink for IC1

Fig. 5 : Details of modulator sereen
bridges, set VR1 fully clockwise and VR2 to midrange and fit the unit in the case with 6BA screws through the fixing holes. Wire the unit up as shown in Fig. 2. Fix two 6 pin DIN sockets in the front of case connected to points G-Q \& S on the p.c.b. to feed the hand controllers. These sockets may be omitted and the multicore wires fed through suitable holes fitted

with grommets. The hand controllers can be assembled in any convenient small plastic boxes.

Switch the power on, monitor the voltage VP across C4 and adjust VRI to obtain $6 \cdot 5$ volts. If you have a scope or counter fit IC4, power up and adjust L1 for 4 MHz . If not, set the top of the core of Ll approximately 6 mm into the former. Fit the remaining inte-

* components

| Resistors All tw 5\% |  | Potentiometers |  |
| :---: | :---: | :---: | :---: |
|  |  | 1kn Horizontal preset 1 | VR2 |
| $47 \cap$ i | R19 | $4.7 \mathrm{k} \Omega$ Horizontal preset 1 | VR1 |
| 10023 | R17, 29, 30 | $50 \Omega$ Wirewound 1 | V\%3 |
| 27001 | R21 | Semiconducters |  |
| $3 \mathrm{k} \Omega \quad 3$ | R20, 25, 28 |  |  |
| $1.5 k \Omega 1$ | R22 | Diodes |  |
| $2 \cdot 2 \mathrm{n} \Omega 2$ | R1, 24 | 1N4601 2 D1,2 |  |
| $4 \cdot 7 \mathrm{k} \Omega 2$ | R18, 27 | Integrated circuits |  |
| $5.6 \mathrm{k} \Omega \mathrm{l}$ | R26 | 4001 A 1. 1C3 |  |
| 10 kS | R11, 12, 14, 15, 23 |  |  |
|  | R313 |  |  |
| $1 \mathrm{M} \Omega \quad 3$ | R7, 8, 16. | $\begin{array}{lll}7805 \text { regulator } & \text { IC1 } \\ \text { AY-3-8710 } & \text { 1 } & \text { IC2 }\end{array}$ |  |
| 2-2MS 1 | R10 | Transistors |  |
| $4.7 \mathrm{M} \Omega \quad 2$ | R4, 5 | BC108 1 TR3 |  |
| 10Mn 4 | R2, 6, 9a, 9b | BC208 1 TR1 | TR1 |
|  |  | BFY50 $\quad 1$ TR2 |  |
| Capacitors |  | BSX20 1 | TR4 |
| Plate Ceramic |  | Switches |  |
| 3.3 pF ? | C 23 |  |  |
| $5 \sim 6 \mathrm{pF} \quad 1$ | C24 | (Arrow CPM3 Black) 2 | S4, 5 (Biased to centre off) |
| 10 pFF | C17 | Paddie s.p.d.t. |  |
| 22pF 2 | C16 | $\begin{aligned} & \text { (Arrow CPM3 White) } 2 \\ & \text { Push-button s.p. } \end{aligned}$ | S7, 8 (Biased to centre off) |
|  | C21, 22 |  | S3, 6,9 |
|  |  | Push-button s.p. 3 <br> Toggle s.p.d.t.  |  |
| Disc Ceramic |  | Toggle d.p.s.t. (Mains) 1. S1 |  |
| 1000pF 3 | C20, 25, 26 | Miscellaneou* |  |
| $0 \cdot 01 / 2 \mathrm{~F} \quad 1$ | $\begin{aligned} & \mathrm{C7} \\ & \mathrm{C8}, 9,18 \end{aligned}$ |  |  |
| $0.1 \mu \mathrm{~F} \quad 3$ |  | Transformer 12V, 12V 250 mA MT12 (Marshall's) Loudspeaker 24 nich 8 . |  |
|  |  |  |  |
| Polyester$0.22 f 52$ | C2, 3 | Hand control boxes (2) RS Type 509-298 |  |
|  |  |  |  |
|  |  | Miniature multicore cable (9-way) 4metres |  |
| Tantalum |  | Miniature mains cable Indicator tamp 42 V |  |
| $0 \cdot 22 \mu \mathrm{~F} 10 \mathrm{~V}$ | $3 \quad C 5,6,13$ |  |  |
| 4.7 F F 10 V$10 \mu \mathrm{~F} 10 \mathrm{~V}$ | 4 C10, 11, 12, 14 | Knob for volume control |  |
|  | $2 \quad \mathrm{Cl} 9,27$ | 6-way DIN plug and socket (2) Optional for hand contral leads. |  |
| $\begin{aligned} & 10 \mu \mathrm{~F} 10 \mathrm{~V} \\ & 100 \mu \mathrm{FF} 10 \mathrm{~V} \end{aligned}$ | 1 C 4 |  |  |
|  |  | Tos heat sink |  |
| Electrolytic |  | Co-axiaf TV socket surface mounting type. |  |
| $220 \mu \mathrm{~F} 25 \mathrm{~V}$ | 1 C 15 | 28-way d.i.I. socket (1) |  |
| 2200رFF25V | 1 Cl | 14-way d.i.l. socket (2) |  |



View of the internal construction of the main p.c.b. and case

Fig. 6: Printed circult board copper track pattern. Ready drilted boards are available from Readers PCB Services (see page 68)



Fig. 7: Main printed circuit board component placement drawing
grated circuits, connect to the aerial input of the television and switch on. Push the reset button, release it and tune the television until a signal from the games unit is found. Several signals may be found, if so choose the best one. Ll may need to be slightly adjusted. When a good picture has been obtained adjust VR2 for the optimum picture, and fit the top cover to the modulator. Check that all the switches function as required and fit the front panel.

## Fault Finding

If the unit fails to function check all your construction carefully, then:-

1. Check that the voltage across C4 is 6.5 volts.
2. Vary VP over the range 6 to 7 volts by means of VR1, if this does not help reset it to 6.5 volts.

|  |  |
| :--- | :--- |
| IC2 Pin Functions |  |
| 1 | Ground |
| $2,3,18,27,28$ | Video outputs |
| $4,5,6,7,8,9,24$ | Control inputs |
| 10 | Reset |
| 16 | VP $(+6.5 \mathrm{~V})$ |
| 19. | 4 MHz clock input |
| $20,21,23,25,26$ | Sound output |
| 22 | Tank trap select |
| $11,12,13,14,15,17$ | Do not connect |
|  |  |

3. Check with a scope that pin 19 of IC2 has a 4 MHz clock input and pin 16 is at 6.5 volts.
4. Check that composite sync is appearing at IC2 pin 18 , pushing and releasing the reset.
5. Check for composite video at TR3 base and again at VR2 wiper. If it is appearing at VR2 wiper try again to tune the television to the game signal. If you still cannot get a picture, substitute a new transistor for T'R4.
6. If the tanks only go forward under control of the switches you will probably find that one pair of wires to the switches are crossed.

## Four Players

Extra excitement and skill can be introduced into the game by splitting the tank controls between four players, two to each tank. The steering controls are operated by the "drivers" while the tank commanders have control of the firing buttons.
The modifications needed to make this a fourplayer game are very simple, especially if DIN plugs and sockets are used for connecting the control boxes to the main unit. The commander's firing control can be fitted into a simple box wired into the DIN plug. If a duplicate set of controls is not desirable then a permanently fitted "commander's" control can be wired into the game unit with a switch arranged to select either the firing button on the "driver's" box or the button on the "commander's" box.


## Pssse wanna waich?

When I.e.d. and I.c.d, watches first came onto the British market, they cost between $£ 50$ and $£ 100$ and any available in the $£ 20$ range were either advertised on 'Police Five' or came of the back of that proverbial loiry.

We are informed by W.K.F. Electronics, that they have a selection of watches available at very reasonable prices.
First, an l.e.d. watch (left in photograph) that features a rather novel touch sensilive operating pad to display hours and minutes, seconds, day and date, also day and month. Price $£ 10.75$ pius VAT.

Second, the standard l.c.d. watch (centre in photograph), which displays hours and minutes continuously,
press the button for day and month, press again for seconds. The watch also has a separate back-light func.ion. Price $£ 10 \cdot 50$ plus VAT.

Finally the chronograph l.c.d. watch (right in photograph) whilst possessing the standard watch functions of hours, minutes, seconds, day, date and month, also features a lap and continuous stop watch facility, with timing to 100 th of a second. Price £ 18.95 plus VAT.

All the watches are supplied with an adjustable stainless steel bracelet and the two l.c.d, watches are powered by batteries costing only 42p (retail) with an estimated life of twelve months.
W.K.F. Electronics, Fleel. House, Welbeck Street, Whitwell, Workshop, Nolts. Tel: 0909 720-695.


## Vero imteresting

Vero Electronics Ltd. Introduce a new range of cases with their 'Series II Boxes AB 010'.
These boxes are moulded in light grey high-impact polystyrene in two parts. The anodised aluminium front panel supplied with the box is retained befween the two halves, avoiding the need for fixing screws. Slots and bosses are moulded into the interior of the box, so that a cholce of mounting positions, either horizontal or vertical, is avallable for p.c.b.'s
or component decks.
Many of the boxes have a battery compartment which is accessible without dismantling the box.

The standard range consists of fifteen boxes varying from $110 \mathrm{~mm} \times$ $68 \mathrm{~mm} \times 33 \mathrm{~mm}$ to $190 \mathrm{~mm} \times 138 \mathrm{~mm} \times$ 91 mm , and other sizes are available to special order at a minlmum quantity of 100 .

Further details from: Vero Electronics Ltd., Industrial Estate, Chandier's Ford, East/eigh, Hampshire SO5 3ZR. Tel: 0421569911.


## No lick TRC-I

The Polycal TAC-1 is a pocket size, liquid crystal display, travel alarm clock. Measuring $65 \mathrm{~mm} \times 32 \mathrm{~mm} \times$ 11.5 mm , it weighs only 45 g with batteries. The 3 V d.c. power input is provided by two silver oxide or manganese alkaline batteries (type GS-14, A-76 or equivalent). Power consumption is 3 mW max. (with the alarm sounding). Accuracy is claimed to be $\pm 30$ seconds per month at $20^{\circ} \mathrm{C}\left(68^{\circ} \mathrm{F}\right)$. With a separate backlight, the multi-digit liquid crystal displays hours, minutes, clock-working sign and am/pm indication.

Priced at $£ 22 \cdot 50$, which includes VAT and $p$ \& $p$, the TAC- 1 is available from: Tempus, Dept. P.W., 19/21 Fitzroy Street, Cambridge CB1 1EH. Tel: 0223312866.

## Rechargeable iron

A new version of the Engel B. 50 Rechargeable Soldering Iron is now available from Kelgray Products Ltd. and is complete with charger unit. The iron now incorporates a built-in spotlight to illuminate the working area and uses long-life rechargeable nickelcadmium batteries. Providing up to 100 intermittent operations ( 300 continuous use) without recharging, which can be achieved in about 8 hours (overcharging is impossible). A safety switch is fitted to the trigger-switch to prevent accidental operation. The iron heats up to an operating temperature of approx $350^{\circ} \mathrm{C}$ in about 7 seconds; a variety of bits are available.

Designed for recharging from a.c. malns, the B50 comes complete with cleaning pad, protective cover, 2 lighting fittings and screwdriver. A particular advantage of this iron is that no stray eddy currents which might damage a sensitive i.c. are generated when the iron is being used.

Priced at $£ 46 \cdot 50$ which includes p\&p and VAT, the B50 is obtainable from: Kelgray Products Ltd., Kelgray House, Sandy Lane, Crawley Down, West Sussex RH10 4HS. Tel: 0342 715066.


This article describes a simple timer which switches an enlarger on and off for predetermined periods. The instrument can handle exposures up to 32 seconds although this can be easily increased as will be described. The accuracy is better than $10 \%$ which is adequate for black and white printing.

The instrument uses a switch which enables the exposure time to be set by feel rather than by having to peer at some dimly lit switch setting.

## Circuit

The circuit is shown in Fig. 2. The 555 i.c. is connected as a monostable multivibrator and on application of a negative pulse to pin 2 a positive pulse emerges at pin 3. The duration of this pulse is determined by the R-C values of R1 to R8 and Cl, and also by the voltage on pin 5 . This pulse is applied to the gate of the s.c.r. which causes it to conduct for the duration of the pulse. The s.c.r. is essentially being used as a mains switch. If the mains were applied directly to the s.c.r. it would only conduct on the positive-going part of the cycle and the light would only have about one third of its normal intensity. A bridge rectifier is used to rectify the mains supply and thus overcomes this problem.

SW2 applies the pulse to start the timing. Pin 2 is tied to Vcc via R9 to prevent unwanted triggering. Application of a negative pulse via SW1 will prematurely terminate any timing interval.

The length of the timing pulse is proportional to the value of the resistance at pins 6 and 7 . The values Ri to R 8 are chosen to give sequence times of $2,4,6$, $8,12,16,24$ and 32 seconds, within $10 \%$ but the values can be altered to suit the needs of the individual constructor. Every $27 \mathrm{k} \Omega$ added to the resistor chain will give an extra two seconds time. Hence if a 48 second timing period is required, an extra $8 \times 27 \mathrm{k} \Omega$ or $216 \mathrm{k} \Omega$ resistor would be needed to be added in series with R1 to R8. In practice a $220 \mathrm{k} \Omega$ resistor would suffice. The final calibration is made bỳ adjusting VR1.
The unit should consume no more than 8 mA , so the power supply uses a miniature mains transformer, with a full wave rectifying circuit.


The timer is built into a small instrument case with all the controls on the front panel

## Components

Any 555 timer i.c. can be used, e.g. LM555, SN72555 NE555, and any s.c.r. and bridge rectifier that can handle at least 400 V at 500 mA can be used for SCR1 and D1.
SW3 is made up from a Doram Mini-Maka switch. The minimum requirement is one pole eight way. The prototype had an open position at one end of the scale, which gives a timing pulse of almost indefinite length. This is used for leaving the enlarger permanently switched on. The mains transformer need only be capable of supplying between 6 V and 15 V at 50 mA at the secondary.

## Construction

The unit is built on a printed circuit board 105 mm $\times 57 \mathrm{~mm}$. The layout of the copper tracks is shown in Fig. 4 and the component layout in Fig. 3. The unit is housed in a metal instrument case and to complete the front panel a $P W$ overlay is available.

The p.c.b. is supported by two short 4BA screws with a nut between the board and case bottom. Remember that some of the copper tracks carry live


Front panel layout of the darkroom timer. A transparent overlay is available from the PW Editorial Offices which will enable readers to achieve a professional finish to their equipment. The overlay is cut out and placed over the front panel, being held in place by the switch nuts and washers. A coloured background can be used behind the overlay if desired

Inside view of the complete unit showing the posifioning of the fransformer, p.c.b. and switch

Fig. 1: Pin connections for the 555 timer i.c. and the C106D thyristor
$\nabla$


## components

Resistors
All $\frac{1}{4}$ W $2 \%$

| R1 | $27 \mathrm{kS} \Omega$ | R6 | $56 \mathrm{k} \Omega$ |
| :--- | :--- | :--- | :--- |
| R2 | $27 \mathrm{kS} \Omega$ | R7 | $120 \mathrm{ks} \Omega$ |
| R3 | 27 kS 2 | R8 | $100 \mathrm{kS} \Omega$ |
| R4 | 27 ks 2 | R9 | $4.7 \mathrm{k} \Omega$ |
| R5 | $56 \mathrm{k} \Omega$ | R10 | $68 \mathrm{k} \Omega$ |

Potentiometers
VR1 $10 \mathrm{k} \Omega \mathrm{min}$. horizontal skeleton

## Capacitors

| Cl | 47 pF | 16 V |
| :--- | ---: | :--- |
| C 2 | 100 pF | 16 V |

## Semiconductors

[^1]
## Miscellaneous

Transformer
SW3
SWa
SW1 and 2

Printed circuit board
Case
Knob
PW Front Panéa Ovelláy


Fig. 2: Circuit diagram of the darkroom timer

Fig. 3: Component layout dlagram. Care must be taken to avoid solder bridges or other forms of short circuits between tracks or components as mains voltages appear at sevesal points on the board. A piece of self adhesive plastic sheet should be placed on the metal base of the case underneath the p.c.b. to avoid any possibility of component leads shorting to chassis

Fig. 4: Copper track pattern for the darkroom timer p.c.b. Ready drilled boards are available from Reader's PCB Service. (see page 68)

mains voltages and that these should be kept well clear of the supporting bolts.

A shielded miniature mains plug and socket are used to connect the enlarger lamp to the unit and this is fitted into the back of the case. If desired the fuse can be a panel mounting type and also fitted into the back. A mains switch is fitted to the end of
the wafer switch. Resistors R1 to R8 are soldered directly on SW3.

Switch to the maximum time period of 32 seconds, then depress SW2 and measure the period. Adjust VRl until a 32 second period is obtained and your darkroom is calibrated.

# Soyoumant to pass the R.-A.E.[Radio Amateurs'Examination] 2 <br> <br> John Thornton Lawrence GW3JGA \& Ken Mc Coy GW8CMY 

 <br> <br> John Thornton Lawrence GW3JGA \& Ken Mc Coy GW8CMY}

Here is the final section of this series and this month we are to consider aerials, feeders, matching and also interference. At the conclusion, we give some hints on actually sitting the examination.

## AERIALS, TRANSMISSION LINES AND MATCHING

The subject of aerials is a complex one and in the space available we will be confining our attention to basic essentials. For further information please refer to the appropriate section in the RSGB Radio Communication Handbook or the Radio Amateurs' Examination Manual.
The fundamental aerial is a length of wire which is half a wavelength long; this is known as the half-wave dipole and is shown in Fig. 85. The aerial is said to be resonant at a specific frequency which is determined by its length, and the distribution of voltage and current along the wire is known as standing waves.
The ratio of voltage to current varies along the conductor, but at the centre of a resonant half-wave dipole it gives a convenient impedance of approximately 70 ohms. If the aerial is broken here, the r.f. power can be fed into the dipole at its resonant point.


Halii-wavelength


Full - wavelength


The full-wave resonant aerial has, as might be expected, a standing-wave pattern similar to two halfwave aerials joined end-to-end. The centre impedance in this instance is very high; inconveniently so in fact and special matching arrangements are called for, as we shall see later.
The radiation pattern of a half-wave dipole is in the form of a "doughnut" shape which in section becomes the characteristic figure " 8 " shape, as shown in Fig. 85.

With a full-wave resonant aerial, the radiation patterns of the two half-wave sections affect one another, producing the four-lobe shape shown.

## Dipole Length

The length of a haif-wavelength in free space is given by:

$$
\mathbf{L}=\frac{300 \times 0.5}{\text { Frequency }(\mathrm{MHz})} \text { metres }
$$

However in practical application, due to (a) capacitance effects at the ends, (b) the velocity of the radio wave being slower in the wire than in free space and (c) the effect of the wire diameter, it has been found that the actual aerial dimensions are about 5 per cent shorter than the calculated free-space length. (Freespace length $\times 0.95$.)
For example, the length of a practical, resonant, half-wave dipole for 3.6 MHz would be given by:

$$
\mathrm{L}=\frac{300 \times 0.5 \times 0.95}{3.6 \mathrm{MHz}}=\frac{142.5}{3.6}=39.6 \mathrm{metres}
$$

## The Vertical Aerial (Quarter wave $\lambda / 4)$

When looking at the radiation characteristics of a quarter-wave vertical aerial, it is necessary to take into consideration the reflective properties of the ground. If we consider the ground as a mirror to the radiation from an aerial, it will be seen that the vertical aerial AB, in Fig. 86 has an image BC in the ground mirror (just as in optics).
Thus, radiation leaving the aerial from point $D$ will travel by two paths in the direction E : one direct from $D$ and the other from the ground reflection. (The position $\mathbf{F}$ is a mirror-image of the aerial in the ground). The radiation pattern is similar to the halfwave dipole but being in the vertical plane it is omnidirectional in the horizontal or plan view.

Vertical aerials fitted on the roofs of vehicles for v.h.f. and u.h.f. utilise the excellent reflective properties of the metal as ground.

## Directional Aerials

The pattern and direction of the maximum radiation of an aerial can be modified by the addition of extra elements, which may be driven by feeding

Fig. 85 : Dipole characteristics


Fig. 86: The $\frac{1}{4}$-wave vertical aerial
power to them or parasitic, where no direct electrical connection is made.
The "Yagi" array shown in Fig. 87 has a half-wave dipole with parasitic director and reflector elements. The Iengths and spacings are chosen to give increased "gain" in the forward direction and reduced gain in the reverse direction (as compared with a plain dipole).


Fig. 87: S1mple directional aerial arrays

One of the consequences of adding parasitic .elements is that the dipole impedance becomes inconveniently low (about 20 ohms) and to overcome this a folded dipole is often used. This has the effect of transforming the impedance up by a factor of four to give a value of around 80 ohms.

## Transmission Lines and Feeders

The source of r.f. power is quite often not the place of utilisation. For convenience we need to have the transmitter indoors, but the aerial has to be outside, as high and far away from buildings as possible. In some instances it may be possible to bring the aerial directly to the transmitter but in most cases a transmission line or feeder cable is required.

## Impedance Matching

For maximum power transfer from one circuit to another the input impedance of the circuit receiving power must equal the output impedance of the circuit delivering it. The output impedance of a valve-type transmitter final amplifier is in the order of a few thousand ohms and a transistorised version would present about five ohms or less. The aerial impedance required can therefore vary between about twenty ohms and several thousand ohms, depending on the type and the point of connection.

The impedance of the transmission line or coaxial feed cable connecting the transmitter to the aerial is defined by its physical construction. Usual values for coaxial cables are 50 or 75 ohms and for twin transmission lines, 70 to 600 ohms, depending on the method of manufacture.

Some form of matching arrangement is therefore required between the various sections of the system which convey r.f. power from transmitter output stage to derrial. A typical example is shown in Fig. 88.

There are three basic types of lines or feeders. (a) Single wire feeder (which carries a true travelling wave.)
(b) Coaxial feeder.
(c) Parallel wire line.


Fig. 88 : Transmission lines

Single feeders (a) are not commonly used because they tend to act as radiators themselves. In the coaxial type (b) the r.f. field is restricted to the inside of the structure, whilst in parallel wires the field is confined to the immediate vicinity of the conductors.

## Characteristic Impedance ( $\mathbf{Z}_{0}$ )

A transmission line or feeder may be considered as consisting of a distributed inductance with associated distributed capacitance, as shown in Fig. 88. It is the relative value of inductance and capacitance which gives the transmission line a property known as characteristic impedance ( $\mathrm{Z}_{0}$ ). When a transmission line is connected to, or terminated with, a pure resistance which is equal to the characteristic impedance, a current travelling along it does not see any change in conditions when it meets the Ioad. In other words, a short transmission line terminated in a purely resistive load equal to the characteristic impedance of the line, acts as though it were of infinite length. Such a line is said to be matched, and here power travels outwards from the r.f. source until it reaches the load, where it is completely absorbed. Let us look at what happens if the transmission line is terminated by its characteristic impedance and then by an impedance other than $\mathrm{Z}_{0}$. This is shown diagramatically in Fig. 89.
Where the line is terminated in its characteristic impedance ( $Z_{0}$ ) the voltage or current will have the same value at any point along it. (a) If however it is terminated with (b) an open circuit or (c) a short circuit, then standing waves are produced along the feeder as shown. This is because the power is not being absorbed at the end of the line but is being reflected: the refiected wave adds to the incoming wave and produces a standing-wave pattern along the line. These examples are extreme cases, but any mismatch produces a resultant standing-wave pattern.


Fig. 89: Transmission line terminations


Fig. 90: A simple standing wave ratio meter

The ratio of the maximum value of the standing-wave to the minimum is known as the standing-wave ratio (s.w.r.). Values will vary from unity (matched condition) to infinity (complete mis-termination).

## Standing Wave Ratio Meter

A useful device for looking at the s.w.r. in a coaxial feeder cable is shown in Fig. 90. Loops of wire, (a) and (b), sample forward and reverse power passing through the centre conductor (c). The voltages developed in the coupling loops are rectified by Dl and D2 and the resulting d.c. output deflects the meter M1, thereby giving an indication of the forward and reverse (reflected) power.

The s.w.r. meter is particularly useful when making adjustments to aerial matching and tuning. Constructional details for a v.h.f. unit were carried in the May 1978 issue of $P W$.

## Matching

Most transmitter and aerial matching circuits are of the resonant type and ane tuned to the operating frequency. We have already described the "Pi" matching network for a valve output stage (PW March 1978, p. 821), shown again in Fig. 91(a): an "L" type network for transistorised output stages is shown in Fig. 91(b). This configuration allows for more convenient component values when working at the low output impedances encountered.

In both circuits the impedance transformation is adjusted by the relative capacitances of C 1 and C 2 , whilst maintaining resonance at the operating frequency. It is usual in this instance to arrange for the transmitter network to provide an output impedance which matches the characteristic impedance of a readily available type of coaxial cable, e.g. 50 ohms or 75 ohms.
When the coaxial cable is operated with a low s.w.r., losses within it are also low, so it is very convenient to fit any filters necessary here.
Some aerials, such as the dipole, have a characteristic impedance at the feed point which will match directly the characteristic impedance of the feeder cable and an aerial matching network is therefore unnecessary. However, if a symmetrical or balanced aerial such as the dipole is fed by coaxial cable, a


Fig. 91 : Transmitter outpuf matching networks
state of imbalance will exist, because one arm of the dipole is connected to the oentre conductor whilst the other is connected to the outer shield. The currents flowing in the shield cannot be cancelled by those in the centre conductor which it surrounds.

## Balance to Imbalance Transformer (Balun)

Diagrams of balun transformers are shown in Fig. 92. In (a) a quarter-wavelength coaxial sleeve surrounds the coaxial cable and in (b) a quarter-wavelength of rod, forming a "stub", balances the output to the aerial. For low frequencies, it is more convenient to wind the balun transformer on a ferrite ring. This type is less frequency-conscious and may be used over a wide range.


Fig. 92: Balun transformers

## Quarter Wave Transformer

Where it is necessary to transform an aerial impedance to match a particular feeder cable, use can be made of a quarter wave "stub" as shown in Fig. 93. Here, a full-wave aerial is to be fed in the centre (where the impedance is around 5,000 ohms) with twin feeder whose characteristic impedance is 72 ohms. If the quarter-wave stub is made to have the correct characteristic impedance then the aerial impedance is transformed down by the stub to match that of the feeder.

$$
\begin{aligned}
\mathrm{Z}_{\mathrm{m}}(\text { matcting stub }) & =\sqrt{\mathrm{Z}_{\text {arrial }} \times \mathrm{Z}_{\mathrm{ilar}}} \\
& =\sqrt{50 \overline{0} 0 \times 72} \\
& =600 \mathrm{ohms}
\end{aligned}
$$

(An open-wire line of 16 s.w.g. conductors spaced 112 mm ( $4^{1}{ }_{2} \mathrm{in}$ ) apart would have a $\mathrm{Z}_{0}$ of 600 ohms ).


Fig. 93: A $\frac{1}{4}$-wave transtorme

## Interference

Non-interference with other radio users, whether they be military, commercial, amateur or domestic, is a condition of the Licence.

An understanding of the way in which interference is caused and how it can be avoided or cured is needed, not only for the RAE, but later on, when you obtain your licence; you will then be in a position to maintain a good clean transmission and live in peace with your neighbours (and the Home Office Inspector).

No practical transmitter is absolutely perfect and in addition to its correct output, is bound to radiate some spurious signals, however small. If these are not kept to a very low level, interference with receivers (TV or radio) operating nearby may result.

Similarly, no practical receiver is absolutely perfect, so when it is tuned to a particular frequency it may be subjected to interference by strong signals on other frequencies, as may be the case if it is situated in close proximity to a radio transmitter.
Interference can also be caused to audio systems, etc., when subjected to strong r.f. fields. Here, the signal enters the equipment and is then rectified or amplitude demodulated, usually by the emitter-base junction of a transistor in the audio pre-amplifier stages, resulting in breakthrough. An increasing number of hi-fi systems employing transistors are prone to interference of this nature.


Al? Capacitors 750 V wkg


Mains Supply Filters Type (b) Using Ferrite Rod


Fig. 94: Power supply filters

## TVI, BCI, AND AFI

Interference can usually be separated into three main categories: television (t.v.i.), radio broadcast (b.c.i.), and audio (a.f.i.).

Television and radio broadcasting are "protected" services and the Post Office may be called upon to investigate cases of interference with these and other authorised transmissions. Audio amplifiers, on the other hand, are not intended to be radio receivers and so will not be afforded the same facilities.

In general, all interference results either from deficiencies in the transmitter or the apparatus being interfered with. Let us look initially at the transmitting end.

## DEFICIENCIES AT THE TRANSMITTER

## Design and Construction

It is important that the various r.f. signals present. within the transmitter are not allowed to radiate directly. Efficient screening is essential, as is the filtering of h.t. and other power supplies, particularly the mains input. A suitable mains filter is shown in Fig. 94. Decoupling and bypass capacitors should be of mica or ceramic, having low inductance properties. (See section on capacitors.) Wiring should be short and direct to minimise stray inductance and capacitance.


Fig. 95: Block diagram of a well-screened transmitting sfation

Tuning capacitor spindles protruding through front panels are often a source of spurious signals and should therefore be of an insulated material or have an insulated coupling.
The cut-out for a panel meter can often cause problems and a screening can over the rear of the meter is desirable. In general, try to ensure that the case of the transmitter is radiation-proof. Commercially made transmitters, including those for the amateur market, already incorporate most of these features and the maker's data sheet usually quotes the level of spurious emissions one may expect.
The block diagram of a well-screened transmitting station is shown in Fig. 95. The transmitter is well protected and its supply leads filtered, ensuring negligible direct radiation. The output passes via a coaxial cable to a harmonic trap, which usually consists of a series-resonant circuit, housed in a screened box and tuned to the particular harmonic likely to cause interference. For example, it might be tuned to 42 MHz , attenuating the 3 rd harmonic of 14 MHz which could be the source of t.v.i. in a Channel 1 reception area. The output then passes via a coaxial cable to the standing-wave ratio (s.w.r.) bridge, which indicates relative forward and reflected power levels.
From here it is fed through a filter, again housed in a screened box which, in the case of a transmitter operating on bands up to 30 MHz , would be of the low-pass type, attenuating spurious signals above this frequency. For a v.h.f. (2m) transmitter, a band-pass filter attenuating spurious signals either side of the pass-band frequencies, would be used.
In practice, the transmitter tuning would first be adjusted into a dummy load. The output would then be switched to an aerial tuning unit which is used to provide optimum matching to the aerial with the minimum of reflected power (indicated on the s.w.r. bridge).
Note: All interconnecting coaxial cables and terminations should be well soldered.

## Aerial and Feeder

The aerial should be sited as high as possible away from neighbouring buildings, TV and radio aerials, etc.
A vertical transmitting aerial is more likely to induce strong fields into nearby equipment than a horizontal one. This is due to the fact that it relies on a ground connection which can cause interfering currents in nearby conductors. In addition, vertically polarised signals are much more likely to be picked up by vertical down-leads, such as those used for television aerials.
It is important that all the transmitter power should be radiated by the aerial proper and that no emission should take place from the feeder cable itself. This means that the currents in each conductor of the feeder should be equal and opposite.

Where a dipole aerial is fed by an unbalanoed coaxial cable, there is significant imbalance in the current distribution and some radiation from the feeder results. The feeder, usually being vertical, readily causes interfering currents to be induced into nearby television down-leads. To overcome this problem, a balance-to-unbalance transformer (balun) is connected at the centre of the dipole, as shown in Fig. 92. In other types of aerials and feeders, correct adjustment of the tuning unit is all-important in reducing feeder radiation to a minimum.

## Operation

Excessive drive in any of the transmitter stages will increase the level of harmonics, so power should be kept to the minimum consistent with efficient operation.

Tuning of the final power amplifier and adjustment of the aerial tuning unit will have a considerable effect on the amount of spurious signals radiated. When tuning the transmitter power amplifier into a dummy load, increase the coupling only until the correct power level is obtained. Do not overcouple the transmitter or the $Q$ of the p.a. tank circuit will be reduced, with a consequent increase of spurious emissions. This also applies when adjusting the aerial toning unit.

An abrupt keying characteristic causes excessive side frequencies, so check each side of your transmission for key clicks (see page 819).
Overmodulation produces excessively wide sidebands and causes splatter; always monitor the modulation level and ensure that overmodulation does not occur (see page 925 ).
The audio bandwidth necessary for good speech communication is about 3 kHz . The modulation circuit of the transmitter should therefore have a rapidly falling response above 3 kHz in order to avoid the radiation of excessive and unnecessary sidebands.

## Summary

Let us summarise the requirements for keeping deficiencies at the transmitter to a minimum:

1. Use correct components in the transmitter, well laid out.


Fig. 96: TV Interference rejection filters
2. Prevent direct radiation from the transmitter and associated leads by screening and filtering.
3. Use appropriate filters in the transmitter output.
4. Use a dummy load for tuning up and a suitable acrial tuning unit, Do not overcouple.
5. Keep aerials in the clear and avoid radiation from the feeder cable (balun transformer).
6. Tune up carefully: do not overdrive or overmodulate.
Check your transmission regularly.

## DEFICIENCIES AT THE RECEIVER

The latest statistics (for 1976) indicate a dramatic fall in the number of complaints regarding transmitter interference. However, the greater proportion of those made were attributed to deficiencies in receiver design.
In many instances, interference is the result of a receiving installation of poor standard; e.g. indoor aerial, aerial incorrect type for area, downlead incorrectly installed, receiver incorrectly installed, excessively long mains leads or speaker leads, etc, etc.

Considering the problem of t.v.i., strong signals can enter the receiver via the aerial and cause interference by cross-modulation in the r.f. or subsequeat stages. A high-pass or rejection filter for the frequency concemed must be fitted in the aerial lead, as shown in Fig. 96: the series-tuned filter being generally more effective.
Masthead amplifiers are a notorious cause of interference as they have broadband input characteristics, some extending from 10 MHz to $1,000 \mathrm{MHz}$. Crossmodulation and swamp effects are common. A highpass filter should be fitted between the aerial and the input to the amplifier, but in practice difficulties arise here because the aerial has to be taken down and the filter made weatherproof.

However, the most common method by which r.f. will enter the receiver is by the presence of "braid" currents in the aerial downlead. These r.f. carrents fiow through earthy parts of the receiver causing r.f. voltages to be produced in susceptible parts of the circuit.

A "braid breaker", suitable for u.h.f. television, is shown in Fig. 97. The reactance of the series capacitors is high at frequencies up to 30 MHz , effectively "breaking" the downlead, but at u.h.f. it is low, resulting in negligible attenuation of the television signal.

Where "braid currents" in the downlead cause interference at h.f. and v.h.f., an alternative circuit can be used. Here a short length of the coaxial downlead cable is wound on a ferrite ring, increasing the inductive reactance of the outside braid without affecting the signals within. An alternative to the ferrite ring, and almost as effective, is to wind the co-ax around a ferrite aerial rod.

If the interference is entering the receiver by way of the mains lead, a mains filter as shown in Fig. 94 should be installed. In the case of hi-fi systems, it can also be picked up on speaker leads, so decoupling these with a disc ceramic capacitor of $\ln F$ to 10 nF is often effective. A ferrite ring may be required in addition, if the problem is really severe.

It is unwise to incorporate modifications inside the receiver, as you may invalidate its warranty and be held responsible for any subsequent malfunction. In difficult cases it would be wise to consult the dealer or manufacturer.


Ferrite Ring Filter (The i.v. feeder is made into a cooxial choke) AD13] Alsa useful on mains \& loudspeaker leads
Ftg. 97: "Braid-breakers" for u.h.f. and h.f./v.h.f.

## Summary

1. Check that the receiving installation is of adequate standard for the reception area.
2. If the interference is entering via the aerial lead, fit a high-pass or rejection filter.
3. If cross-modulation occurs in a mast-head amplifier, fit a high-pass filter in the aerial feed to it.
4. If the interference is entering via the downlead braid, fit a ferrite ring or capacitive braid-breaker.
5. If the interference is entering via the mains lead or other cables, fit a mains filter or ferrite ring.
6. If the problems are caused by direct radiation, try repositioning the aerials, feeders, etc. Wherever possible, avoid making internal modifications to receivers or audio systems. In difficult cases, refer to manufacturer.

## Note:

An excellent series of articles on interference appeared in Radio Communication, May 1975. Some back-issues are available from the Radio Society of Great Britain, 35 Doughty Street, London WCIN 2AE. Alternatively your local Radio Club or a nearby radio amateur may be able to help with a copy.

## SITTING THE R.A.E.

It is a good plan to set yourself regular revision two or three nights a week prior to the examination. Allocate specific subjects each night so that everything is covered in good time and practice answering specimen questions in writing and time yourself.
It is best to leave the night before the exam relatively free from commitments. This is a good time for collecting together the things you need, including your thoughts!

Take with you, two pencils and a pencil sharpener, two pens (in case one runs out), ruler, symbol stencil (if you have one), eraser and your examination card. Calculators are permitted provided they are of the electronic, battery-operated type. There are no restrictions on the functions the machine will perform.

You should receive your card at least a week before the examination. If not, contact the Examination Centre/Night School immediately and confirm that you are, in fact, registered.

When going to the Examination Centre, allow plenty of time; your bus may be delayed, you may have difficulty in parking your car. Aim to be in the building at least 30 minutes before the start. Locate the examination room, and be seated well before the examination commences.

When appropriate, your invigilator will distribute the papers: do not look at them until he gives the signal. Timing will begin from the point at which you are allowed to turn the paper over-don't panic! Give yourself at least ten minutes to read things through carefully and note any special instructions.

If, through no fault of your own, you arrive late, apologise to the invigilator then settle into your place quietly and attempt as many of the questions as you are able in the period remaining.
Remember, the paper is divided into two parts; Part 1 has two compulsory questions on licensing conditions ( 15 marks) and interference ( 15 marks). Part 2 requires six questions from a choice of eight to be answered ( 10 marks each). Failure in either section will regretfully result in failure of the examination as a whole.
Give yourself about twenty minutes for each of the eight questions and this will allow a little time at the end to look through your answers. The answers should not be overelaborate or padded out: make your point as quickly and as clearly as possible, there is no time to waste. It is probably best to tackle the compulsory questions in Part 1 first and then move on to Part 2 where you have a choice. Here you can sonser the easiest first and so gain a little extra time.

With answers containing calculations, it is a good idea to set out all the steps in detail. Then if a slip is made in the arithmetic, the examiner will see the correct formulae and methods have been used, and will mark accordingly.

Hints and tips on the drawing of diagrams were given earlier in this series (February 1978, page 764), but remember drawings take time, so keep them simple. You are not being examined in grammar or indeed spelling but neatness and cleanliness are relatively important, if only to avoid ambiguity. The examiner will not be impressed by illegible writing, scribbled diagrams, etc. If you make a mistake, cross it out with a single ruled line; it will then be ignored and you will not lose marks.

A few minutes before the end, stop, read through your answers and correct any minor omissions. Make sure all are numbered and lettered correctly and that your examination reference is quoted.

We wish you every success and hope to meet some of you on the air in the near future.

73 de GW3JGA es GW8CMY.

## NOTE

The syllabus and examination pattern for the R.A.E. are changing in 9979. We plan to publish a follow-up article to update this series


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All those functions, from just 10 keys? In such a small calculator? The secret lies in the special four-level keyboard. Each leve! has a different set of functions. Simple two-way switching system allows you to select any keyboard level quickly and easily. Each set of functions is carefully grouped, to let you whisk through calculations with the minimum of switching

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## Science

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by Eric Dowdeswel/ G4AR

Our SSTV expert Paul Barker of Sunderland has now become G80VD and soon will be able to look back at the other chaps! Paul studied at home for the RAE and had only the RSGB's RAE manual and a few library books to see him through. He managed to "copy" three VKs on 20 m SSTV so completing his SAC! Best of luck on the air OM. Likewise John Overton did the trick in Milngavie, Glasgow and so sports GM4GUA now, having already passed the code test. He's using an HW32A on 20 m s.s.b. for the time being but a KW2000B is on the way.

Brian Smith is well away in Barry, Glam, with his new FRG7. Having been disgusted with what he has heard of the CB band just below 28 MHz he says he'll get stuck into the RAE studies and get a proper ticket! After a spell of ten years away from the amateur bands Ken Proctor of 35 Hertford Close, Eastield, Scarborough, Yorks, is active again and would like to hear from old friends including Bernard Hughes BRS25901 who contributes to this column. Ken has a Trio 9R59DS and 90ft of wire and covers all the bands.
The note on David Greenhalgh in the February column resulted in G3SHW of stockport coming up with a Lafayette HA350 for him which illustrates a little of the fine spirit of amateur radio. David, in Poynton, Cheshire is delighted and hopefully he'll be contributing to the column in future.

After using a domestic portable for a while J. Gooditer of Marple, near Stockport, acquired an FRG7 and really found out what the amateur bands are all about. He wants a good prefix list and so, as ever, the answer is Geoff Watts of 62 Belmore Road, Norwich who will deliver the goods for just 40p. From Newport, Gwent comes a first report from Martin Yiezers and a note on his local ARS, reported later. He has 250 ft of wire on his Realistic DXI60 and an 8 ft rod aerial for "the h.f. bands". I've suggested that he try the long wire on those bands too!

Now here is an interesting letter from Bill Land BRS34761, 7 Wellbrook Road, Bishops Clare, Cheltenharn, Glos, who admits to being 72 years young and an addict of RTTY for which he has a Creed 7B printer running from his Eddystone 750 and Trio 9R59DS. Bill wants to hear from others using this mode and can help with a list of parts for the 7B. Another old-timer is John Whiting of Fareham, Hants, who started with crystal sets in 1922 but now sports an FRG7, listening mainly on 10 m for the moment. He can cope with the code test so is contemplating studying for the RAE.

Fame at last! Bob Griffiths writes for the first time to say we have a keen young reader on the Isle of Wight! He, too, has had a lay-off of 14 years but is now back in the swim again but with a lot of pertinent remarks on the plethora of UA stations! Must come as a shock after the relative hush of the bands in days gone by. Bob uses an EA12, AR88D plus converters for 2 m and 70 cm . His aerial farm consists of 132 ft wire, 20 m ground-plane, 20 m V-dipole and a vertical for 80 m .

From Morpeth, Northumberland comes John Hodgson again with DX heard on his Realistic DX160 and long wire. He has received QSL cards from CB'ers in Sweden and Norway! John is keen on RTTY so suggest he writes to Bill Land mentioned above. Bob Bell of 5 Byron Avenue, Blyth, Northumberland can't be too far from John just quoted above and they ought to get together since Bob would like to start a club in the area. Bob is yet another FRG7 owner, having been SWLing for some 25 years and worked his way through a number of sets.
CLUB NEWS Geoff Cole G4EMN, secretary of the Wessex ARG, oft mentioned here, had applications for membership from a couple of readers situated a long way from Wessex. Although flattered, he says that this is not quite the ideal So, anyone who wants to join a club, try to locate your local group or write to the RSGB, 35 Doughty Street, London WCl for the address. Geoff rightly says that every reader of this column ought to belong to a club.
Blackwood and District ARS GW6GW meets every Friday 1930 hrs at Oakdale Community College, near Blackwood, Gwent. On May 12th GW8LJJ talks on Practical Construction and the 19th is film night, Printed Circuit Manufacture. Write to Steve Cole GW4BLE, 10 Llanthewy Road, Newport.
Mid-Warwickshire ARS first and third Mondays 2000 hrs at 61 Emscote Road, Warwick with G3UDN on the air. They'd like to see PW readers, particularly Nick Smith A9050 and C. J. Roe of this column.

Contact Norman Read G8CXL, 86 Telford Avenue, Leamington Spa.

Bury RS G3BRS, every Tuesday at Mosses Centre, Bury at 1930 hrs with RAE courses, code training, local c.w. nets, new local repeater GB3MA on 70 cm . Whew! It's all go! Hon. Sec. is Eric Thirkell G4FQE, Mosses Community Centre, Cecil Street, Bury.

Stevenage and District ARS get together at 2015hrs on first and third Thursdays at Staff Canteen, Hawker Siddeley Dynamics. Morse classes and RAE course at local college by G3SJR. May 18th is the date for a lecture by members of London UKFM Group, while on June 1st G3AGP talks on Electronics in Medicine. Contact Trevor Tugwell G8KMV, 11 The Dell, Stevenage, Herts.

Silverthorn RC produces a ten-page newsletter "Spurious" so they're pretty active, meeting every Friday 1930 hrs at Friday Hill House, Simmons Lane, Chingford, London E4. Contact C. J. Hoare, Hon. Sec., at this QTH.

Newport ARS meets Mondays 1930hrs at Brynglas House Community Centre, Newport with RAE and code classes coming up. Write to Martin Liezers, 32 Barrack Hill, Newport.

## Log extracts

B. Smith:-80m C5ABC (Gambia) 4U1ITU 20m C5AAF FC9UC JY5HH ZL2AM $15 m$ CE3BPC C31LU
K. Proctor:-80m EA8QJ ZB2G PJ8CO KL7AVX ZL3BX CO7RS 9Y4NP 40m CO4DC HP1XYA ZL1BAQ ZL2BDP ZL3RD 20m VP8PT HP3XKB HL1JI HK0BBF 55m FP8DX KL7JEJ JA9IWN 10 m FG7BA
J. Hodgson:-80m J3AAG OY5NS PJ8CO 8P6FV 40 m TG4NX YV5APF 20m FM7WV HK1NR VE8RCF VK7AE VP2DAW 15m HK0QA KL7GRP ZS2MI (Marion Is) ZB2G 10m EA9FL HH2MC VU2DK.
J. Whiting:-10m CT2AX EA8BS SV1HX.
B. Land:-RTTY 20m DK6RY EA3ABU IT9BWT JY5KR K6XP OE2KO SM6HUG YU2BOR W2WIX. M. Liezers:-80m KZ5JM ZB2G 5B4DI 20m HK0QA KM6FC YB0HH 15m KL7GRP VP1AB 10m C5ABC EP2RL.
J. Goudier:-20m CT4YN FC9UC.

All s.s.b. except where stated otherwise.


by Ron Ham BRS15744

Richard Staples, G8MME, Lymn, Cheshire, is active on both 2 m and 70 cm s.s.b. On 2 m he runs 400 watts from a pair of 4250 A valves to a couple of 10 -element parabeams, at $40 \mathrm{ft} \mathrm{a} . \mathrm{g} .1$, and on 70 cm , his 2 m exciter, TS700G, drives a Microwave Modules Transverter and a home-brew 4CX250B amplifier producing 250 watts to a 48 -element multibeam, mounted above his 2 m array. Richard is looking for s.s.b. skeds on 70 cm and will be reporting on future activity in the north-west of England. Vic Hartopp, G8COB, Northampton, uses an IC201 transceiver for all modes on 2 m and has

> Reports on the various bands are welcome and should be sent direct, by the $45 t h$ of the month, io:-
> AMATEUR BANDS Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey KT 21 gTW. Logs by bands, each in alphabetical order.
> MEDIUM and SW BANDS Charles Molloy G8BUS, 132 Segars Lane, Southport, PR8 $3 J G$. Reports for both bands must be kept separate.
> VHF EANDS Ron Ham BRS15744, Faraday, Greytrlars, Storrington, Sussex RH20 4HE.
recently been enjoying himself rebuilding a Mohican receiver, and listening to the American CBers on his HRO.

A couple of newcomers to v.h.f., Henry Hatch, G2CBB, South Croydon, using an IC215 is working through the repeaters and Robin Bellerby, G3ZYE, Hove, Sussex, taking part in his first 2 m opening, did not realise there could be such pile-ups of stations on the repeaters. In Littlehampton, Sussex, Norman Langridge, using the v.h.f. section of his Yaesu FR400Super DX receiver with a 3-element home-brew, loft mounted beam can always hear traffic through GB3SN, but, during the opening on March 7th he heard GW mobiles working through GB3BC which has encouraged Norman to modify his aerial systern. Along the coast to Worthing where David Wakefield, BRS39756, uses the receiver of a Pye Vanguard on 2 m , with a roof mounted dipole, and CR100 and R1475 receivers, fed by a long wire aerial, on the h.f. bands. David's interest in radio is further stimulated through his activities as a Cadet Wireless Operator in the Air Training Corps.

Despite the poor conditions and the low atmospheric pressure, 29•3in, Alan Baker, G4GNX, Newhaven, had a 59 contact on 2 m s.s.b. at 1800 on February 26th with F1ENH/P, Boulogne, both running 15 watts, and at 1730 on March 2nd, in thick fog, on Beachy Head, near Brighton, Alan heard signals from repeaters ONOHT and ONOOV and a Buckingham station working through the Kent repeater, GB3KR, $R 4$.

The atmospheric pressure remained around $29 \cdot 5$ in from midnight on February 27th to midnight on March 2nd when it took off and climbed to $30 \cdot 4$ in by noon on the 6th, and, true to form, the v.h.f.s opened up as the high pressure began to fall. During the evening of the 6th, John Kuipers, G4GUX, Brighton, worked GBLY, in Hampshire, on 70 cm , first via GB3AW, Ashmansworth, Berks, RB10, and then through GB3BR, Brighton, RB6, and GB3PH, Portsmouth, RB2. Later he heard signals through GB3BK, Upper Basildon, RB6, GB3SD, Weymouth, RB14, had a c.w. contact with ON4VN and heard 2 m signals through GB3BC, R6, GB3MH, on its new channel, R3, GB3WW, Carmel, Dyfed, R7 and FZ3THF, R4.

Around 1945 on the 6th, Ken Smith, BRS 20001, Horsham, received several French broadcast stations in Band II and patterning on u.h.f. TV channels, and earlier, Graham Laucht, G80QM, Birmingham, worked G8KSN, Ramsgate, Kent first via GB3PI, Cambridge, R6, and then direct. Like Graham, Brian Fenwick, G8BTC, Brighton, heard a variety of repeater signals on 2 m , including Cornwall, GB3NC, R5; Brian noticed that signals were often stronger on the input frequency than from the repeater. Frequently, on the

6th and 7th, I received a good picture from Lichfield, Ch 8, 189 MHz , Continental f.m. stations in Band II and strong signals from GB3BC, BM and KR, with only dipoles feeding the respective receivers. At 0125 on the 7th, I heard two Portsmouth stations, G8NUI and G80QN/M in contact via the London repeater, GB3LO, R7, and at 2152, another pair, G4GUX and Martin Newell, G8KOE, both in Brighton carried out a similar test on 70 cm and had a QSO via the Portsmouth repeater.

At 0900 on the 7th, G4GNX/M using his new IC-240 to a $5_{8}$ whip aerial, heard strong signals through GB3BC, from a low point in Brighton, F1EBE, Rouen, via GB3LO, and at midday on the 6th, he received signals from the Malvern Hills, GB3MH, R3, and Paris repeaters. At 2206, from Newhaven, Alan had a c.w. contact with ON6FT and at 0011 on the 7th he worked FlEVM/P, Caen, on 2 m s.s.b.

The atmospheric pressure continued falling until 1800 on the 8 th when it began rising rapidly, reaching 30.4 in by midday on the 10 th , and then fell back to $30 \cdot 0$ in by noon on the 12 th which, as expected, caused a tropospheric opening.

At 2230 on the 8 th, newly licensed $6800 \mathrm{M} / \mathrm{M}$, using a FT227R to a ${ }^{1}$ o whip aerial, travelled down to Beachy Head, Sussex, and heard GB3BM, his home repeater and then had QSOs through GB3BC. From Brighton's Race Hill, at 1900 on the 9th, G4GNX/M worked F1BBZ and F3KT via FZ8THF, Vichy, R0, a distance of 320 miles. Roy Bannister, G4GPX, Lancing, Sussex, succeeded in accessing the Vichy repeater but owing to UK traffic on R0 he could not complete a QSO. Both Alan and Roy heard the Ghent, ONOON, R4, and Paris, R6, repeaters and during the evening of the 10th, Clive Penna, G3POI, Orpington, Kent, worked EAlAB on 2 m c.w. and Mike Rowe, G8JVE, East Preston, Sussex, heard an EAl and worked into France and Wales on 2 m s.s.b. Around 2330 on that evening, Alan Belfeld, G4GLN, Streatham, contacted an EA on 70 cm c.w.

John Branegan, now licensed GM80XQ, Saline, Fife, worked his first three countries, G3GZX, GI8JTS, and GW4GTE during the 2 m contest weekend, March 4th and 5th. John said "there was a definite lift on March 4th, and by 2200, GI signals into Scotland were very good and they could hear the Central Scotland repeater in GI".

Throughout the period, 9th to 12th there was frequent co-chanmel interference an Band II and u.h.f.TV and at 0914 on the 11th I received a 599 signal from the Emley Moor beacon, GB3EM, on 70 cm with only a dipole feeding the receiver.

The 10 m band has been open almost daily from February 20th to March 15th with a familiar pattern of strong signals from Russian amateurs in the morning and from north-American amateur and CB stations during the afternoon and early evening. I heard VKs during the early mornings of February 26th and March 7th and JAs on March 5th and 11th working into Europe. On most days signals averaging 539 were heard from the Cyprus beacon, 5 B4CY and the project TESSA beacon, ZE2JV. During the afternoons of March 1st and 2nd, Don Butterworth, G3IKO, Redhill, Surrey, heard the Bermuda beacon VP9BA and I logged the Bahrain beacon, A9XC, during the afternoons of the 5 th and 6th.

February 26th was a good day on 10 m for Gordon Goodyer, BRS 37345, Petworth, Sussex, who heard both A9XC and VP9BA in the morning as well as signals from amateurs in 22 countries. Although
signals were very strong throughout the day Gordon reports a drop off in strength around midday and a quiet band from 1430 to 1515. Roger Bunney, DX TV columnist, using his Hallicrafters $5-10$ receiver logged transatlantic signals up to 37 MHz on the 26 th . Nigel Golds, BRS 36910, West Chiltington, Sussex, noted the powerful American stations around 1730 on March 1lth and the Russians during the afternoon of the 12th.

The exceptional 10 m conditions observed by Gordon and Roger on February 26th were followed by ionospheric disturbances, reported by the BBC World Service, on the 27 th and 28 th which may well have been caused by the solar storm recorded by Cmdr Henry Hatfield, Sevenoaks, Kent, John Smith, Rudgwick, Sussex, and myself from February 21st to 26th. What's more John Branegan, reported auroral activity, with plenty of 2 m c.w. from EI, GW and LA, from 1830 to 1912 on February 27th and from 1657 to 1912 on March Ist. John has built a projection box attachment for his 2 in telescope so that he can record the number of sunspots and since 2048 on March 5th he has been monitoring the telemetry signals from the new amateur radio satellite, OSCAR-8.

Despite frequent overcast skies, Henry Hatfield got a glimpse of the sun through his spectrohelioscope on March 6th when, in addition to sunspots, he saw several filaments and a ribbon flare, on the 10th he counted some 15 spots in four main groups, and on the 15th he noted a bright plage on the east limb. Henry, John Smith and myself recorded solar noise at 136 and 142 MHz on March 2nd, and then daily from the 6 th to the 13 th with noise storm conditions on days 6 th, 8 th, 10 th and 15 th.

According to comments heard both on and off the air by Alan Baker and myself the "Slim Jim" 2m aerial described in our April issue works well and is very popular.

Let's hope that conditions will be good to provide plenty of DX for the RSGB's HF National Field Day on June 3rd and 4th and VHF NFD and SWL contest on July Ist and 2nd. Thanks again for all your reports, and good Iuck if you compete in the contests.

## What do the VHFs have to offer?

The accepted part of the radio frequency spectrum known as Very High Frequency ranges from 30 to 300 MHz , ( $10-1$ metres), and is generally used for communications by aircraft ( $108-136 \mathrm{MHz}$ ), amateur radio ( $4 \mathrm{~m}, 70 \cdot 05-70 \cdot 7$ and $2 \mathrm{~m} 144-146 \mathrm{MHz}$ ), private and business mobile radio ( $7 \mathrm{I}-88$ and $165-174 \mathrm{MHz}$ ), shipping ( $156-165 \mathrm{MHz}$ ), satellites $(136-137 \mathrm{MHz})$, military, various emergency services and broadcasting by f.m. radio, Band II ( $88-108 \mathrm{MHz}$ ) and television, Band I ( $41-67 \mathrm{MHz}$ ) and Band III ( $176-215 \mathrm{MHz}$ ).

The effective range of a v.h.f. signal is naturally limited by the terrain beneath its path and the prevailing atmospheric conditions. Broadly speaking, the propagation of signals between $30-80 \mathrm{MHz}$ is governed by the E region of the ionosphere, and above 80 MHz , by conditions in the troposphere. These bands are therefore a challenge to both the listener and the transmitting amateur, to be ready for sudden sporadicE disturbances between April and August and at any time, for a tropospheric opening, when the weather
is fine and the atmospheric pressure is high. These v.h.f. openings are exciting, because the signals travel more than ten times their normal range and as the v.h.f. bands are shared throughout Europe, there is a mine of DX, both sound and vision, to be found among the mix-up of continental signals while the event lasts.

During sporadic-E events, very strong signals are heard in the UK from east-European broadcast stations between 65.73 MHz and even stronger signals, from a variety of continental mobiles, interfere with the Band I television channels. There are two beacons in the 4 m band to listen for, GB3SU, $70 \cdot 695 \mathrm{MHz}$ and GB3SX, $70 \cdot 685 \mathrm{MHz}$.

The 2 m band is full of surprises throughout a tropospheric opening; and like most v.h.f. enthusiasts, our readers have heard and worked amateur stations, using all modes, over a wide area from Scandinavia to the Mediterranean Sea, and all parts of the UK.

First indications of an opening can come from any of the chain of 2 m beacons ranging from Cornwall, GB3CTC, $144 \cdot 915 \mathrm{MHz}$, to Angus, GB3ANG, 144.975 MHz , and northern Ireland, GB3GI, $144 \cdot 137 \mathrm{MHz}$, to Wrotham, Kent, $144 \cdot 150 \mathrm{MHz}$. Having decided, from the number of UK beacons heard, the extent and predominant direction of the prevailing lift, it is worth looking for the continental beacons such as; DLOPR, 144.910 MHz , EA3URE, $144 \cdot 042 \mathrm{MHz}$, FX3THF, 144.905 MHz , LAlVHF, $\quad 144 \cdot 860 \mathrm{MHz}$, OH6VHF, $144 \cdot 900 \mathrm{MHz}$, OK0EB, $144 \cdot 970 \mathrm{MHz}$, ON4VHF, $145 \cdot 990$, OZ7IGY, $144 \cdot 930 \mathrm{MHz}$, PA0JTA, $144 \cdot 148 \mathrm{MHz}$, SK4MPI, $144 \cdot 960 \mathrm{MHz}$, SP2VHF, $144 \cdot 980 \mathrm{MHz}$, and YUIVHF, $145 \cdot 990$ to name a few. A beacon's signal is a continuous note, frequently interrupted with its call-sign.

Another good propagation indicator is the 2 m repeater network, using f.m., which now provides considerable coverage of the UK and is rapidly spreading through Europe. Each repeater shares a carefully planned channel numbered from R0 through R9, covering a series of input frequencies from $145 \cdot 000 \mathrm{MHz}$, in 25 kHz steps, to $145 \cdot 225 \mathrm{MHz}$. The range of output frequencies, also in 25 kHz steps, is 600 kHz higher, from $145 \cdot 600$ to $145 \cdot 825 \mathrm{MHz}$. In addition to handling the amateur radio traffic, these unmanned, automatic duplex transceivers periodically identify themselves by transmitting their official callsigns.

At some time during a tropospheric opening, continental f.m. broadcast stations will be heard in Band II, often stronger than the "Iocal" BBC stations, and Band III television will suffer from co-channel interference.

Space enthusiasts can use the AMSAT-OSCAR 7 satellite which has two repeaters (transponders) aboard, 2 m to 10 m and 70 cm to 2 m . Signals received by the satellite between $145 \cdot 85$ and $145 \cdot 95 \mathrm{MHz}$ are re-radiated between 29.4 and 29.5 MHz and signals going in between $432 \cdot 125$ and $432 \cdot 175 \mathrm{MHz}$ come out, inverted, between 145.975 and 145.925 MHz , The satellite's telemetry beacon can be heard on $145 \cdot 980 \mathrm{MHz}$.

Every year, the RSGB, in conjunction with the IARU, arrange a number of contests, on 4 m and 2 m , for licensed amateurs and SWLs. These events are interesting because they are well supported by individual and group entries operating from their home QTH, or from portable or mobile locations. Should a contest coincide with an atmospheric disturbance then there is plenty of DX about.

Sections of the 2 m band are used for Slow Scan TV, RTTY, the Radio Amateurs Emergency Network (Raynet), and scientifically, for moon-bounce experiments, meteor scatter and the study of auroral propagation.

I suggest that readers who are interested in v.h.f. should obtain a copy of the RSGB VHF/UHF Manual and talk to the v.h.f. operators in their local radio club.


## SHORT WAVE BROADCASTS

by Charles Molloy G8BUS

Harmonics, though familiar enough to the radio amateur, may well be an unknown phenomenon to the broadcast band DXer. A harmonic in this context is a spurious transmission on twice or three times the frequency of a broadcasting station. The basic frequency is called the fundamental, twice that frequency is the second harmonic, three times the frequency is the third harmonic and so on. Harmonics occur naturally inside a radio transmitter but great efforts are made to suppress them so that they do not reach the aerial and radiate. High power transmitters are now commonplace and when connected to a directional array the effective radiated power (e.r.p.) may easily be in megawatts. Under these circumstances it is easy for a few watts of the second or third harmonic to be radiated along with the fundamental. Normally, this is not a problem as the s.w. broadcaster ought to be transmitting near to the maximum usable frequency (m.u.f.) to obtain optimum results. Harmonics would then be above the critical frequency and will penetrate the ionosphere and not come back to the earth.

Now that the higher frequencies are opening up again after the recent sunspot minimum, reception of harmonics should occur more often. Radio amateurs make world-wide contact on 10 metres using only a few watts, and reception of harmonics in the range 23 to 30 MHz should occur over a similar range. Harmonic DXing is the broadcast band DXer's equivalent of QRP (low power) and as such has a challenge and fascination all its own. One snag though. Be sure you are listening to a genuine harmonic and not to a spurious signal generated within your receiver. A check for the fundamental or a check of conditions on 10 metres is a help.
From Guernsey in the Channel Islands comes a letter from George Le Couteur who heard Radio Moscow when he was tuning through the 10 m band. The exact frequency was 28.350 MHz which is three times 9.450 in the 31 m band and sure enough he "found them putting out their usual huge signal on the frequency". The second harmonic is 18.900 and George found them on this out-of-band frequency as well. Other DX heard, on fundamental this time, is Radio New Zealand regularly around 0830 on 11820 kHz , faint but clear of QRM.

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Harold Brodribb, St Leonards-on-Sea has also picked up harmonics of Radio Moscow, on 23.920 MHz ( $2 \times 11960$ ), $24 \cdot 110(2 \times 12055)$ on $29 \cdot 1$ (?) and $29 \cdot 8$ (?). Harold is now using a home-made converter with his CRIOO which gives very much better results in the range $25-30 \mathrm{MHz}$ and he reports increased activity in the 11 m broadcast band ( 25605 to 26095 kHz ) with reception of Tel Aviv on 25605 at midday and of the Voice of America (location unknown) on 26000 kHz at 1800.

Short-wave crystal sets are in the news again with a report from Rod Hunt of Darlington who pulled in Radio Canada, Finland, Norway, Sweden, Austria and a number of others, using a home-brew crystal set and a 50 ft outdoor aerial. Rod says that one definitely needs patience and persistence and it is only a matter of time until Australia is caught!
Radio Australia has attracted the attention of George Norris of Stowmarket (FRG7 and Windom Antenna) who says that propagation on this path at the moment seems to favour the morning between 0700 and 1000 . "Try 11740 kHz ( 25 m ) first and then 21680 and 21570 ( 13 m ). The former seems to hold up well most of the day with QRM at times". William Stevenson (Manchester) has a Vega 206 which he uses with a folded dipole in the loft for the 19 m band and a 45 ft Zepp with a.t.u. for the rest of the bands. His $\log$ included Radio Australia on 11705 kHz at 1500 SIO 434. Fifteen-year-old Robert Pound of Cambourne in Cornwall is a regular listener to Radio Australia on 11740 but has been trying without success on 4920 for the 10 kW VLM4 which relays the Domestic s.w. Service between 1930 and 1400. Has anyone logged this station? Robert, who uses an exWD receiver (type unknown) and an end-fed aerial situated in the loft, has beeen trying to pick up Radio Kaduna, Nigeria on 3396 kHz in order to hear the talking drum interval signal. The NBC international service is also using this interval signal. Chris Howles of Lichfield (Vega 206 and 50ft long wire) reports hearing Lagos signing on at 1800 on 15120 kHz in the 19m band and according to the World Radio and TV Handbook the interval signal precedes signing-on.


## MEDIUM WAVE DX

by Charles Molloy G8BUS

A long interesting letter, just bursting with enthusiasm from Mike Kuske of Folkestone describes how he heard his first North American m.w. station. His National Panasonic GX600M had been purchased for short wave listening but after reading $P W$ he decided to try the medium waves for, although interested in North American domestic radio, he never thought it could be heard in the UK. With 100 ft of wire wound around the loft for an aerial he pulled in WINS in New York City on 1010 kHz at 0045 and he continued
to listen to it until 0400. Subsequent night listening between 0100 and 0330 produced WHN New York on 1050, CHUM Toronto 1050, CFRB Toronto 1010, CHNS Halifax 960, CJYQ St John's 930, WCAU Philadelphia 1210, WOWO Fort Wayne Indiana 1190, CJCH Halifax 920 . Mike now intends to obtain one of the new Yaesu Musen digital receivers and he also plans to construct a loop. A number of points arise out of Mike's letter.
Reception of North American m.w. stations is not limited to the winter. DXing is possible throughout the year. All that is required is a path of darkness between transmitter and receiver and this occurs some five hours after local sunset in the UK. Even on the longest day, sunrise oocurs seven hours after sum set at this QTH though this period may vary according to location in the UK. At my QTH this means about $1_{2}$ hours DXing of NA before dawn on June 21st and much longer periods during April, May, July and August. Only stations on the eastern seaboard of North America will be heard in June but on the other hand European QRM is light as Central and Eastern Europe will be in daylight. DXing North America in summer can be rewarding. I once logged the 40 watt relay at Glovertown in Newfoundland on 1090 kHz during the month of July.
Mike noticed that fade-outs lasting for periods of several days are common on the North American path and he wonders if there is any set pattern for good reception. Fade-outs do recur sometimes after 28 days, which is the period that a disturbance or spot on the surface of the sun takes to rotate and face the earth again-wthat is if it lasts that long. Reception is often at its best just before a fade-out so it might be possible to predict such a peak. Matters are not so simple, though, as the ionosphere is a complex medium and long range propagation on the medium waves has not been studied to any degree. After all the band is intended for looal broadcasting! In any event the medium waves are never dead, there is always some DX to be heard. If North America cannot be heard at all then South America will be at its best. If you hear an American voice during a fadeout then listen carefully for it may be something special such as Puerto Rico or perhaps the American Virgin Islands. I shall give up medium wave DXing if it ever becomes possible to predict what can be heard as most of the fun will then have disappeared.

Finally, Mike offers a tip to QSL hunters. International Reply Coupons cost 25p in the UK but stamp dealers can supply unused postage stamps for most foreign countries. These are more convenient for the recipient to use and, a lot cheaper to buy (for Canada in Mike's experience). A very good idea if you know what value of stamps to obtain for a return letter.

Central and South America are often heard well during the summer. D. R. Mayhew (Littlehampton), who uses a Philips receiver and the PW 40in loop reports hearing Radio Managua, Nicaragua on 620 kHz , La Voz de Mexico on 730 , Radio Cadena el Salvador on 760 , 'ROK 80 ' in Juarez Mexico on 800 , Radio Sutatenza Colombia 810 and Radio la Versatil El Salvador on 1300 kHz . More Latin American DX is reported by Steve Whitt (Cambridge) who used an Eddystone 940 with a 2 ft square loop and the $P W$ balanced amplifier to pull in Radio Margarita 1020 and Radio Coro 1210, both in Venezuela. "It was very interesting to receive my first South American DX, I don't know how I missed it before," writes Steve. It is very easy to pass over Latin Americans, especially during a North American fade-out, and class

## HERE IS THE NEWS

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| 7429 | 48 | 74t43 | ${ }^{45 \mathrm{p}}$ | 742S134 | 149p | 4020 | $1{ }^{40}$ |  |  | 17 | 20ie2 | 4 p | EF23？ | 34p | T1P35 | 14p | 2N4125／6 | 22 p | OA E00Y 120p | 4，30 $T$ | Tog |
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