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# OCTOBER 1977 • VOLUME 53 • NUMBER 6 

## BRITAINS LEADIMG IOURNAL FOR THE RADIO \& ELECTRONIC GONSTRUGTOR

Published by IPC Magazines Ltd., Fleetway House, Farringdon Street, London, EC4A 4AD Telephone: 01-634 4444

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    News and Views
406 EDITORIAL-I Spy!
407 NEWS...NEWS...NEWS
ON THE AIR-Amateur Bands. . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . Eric Dowdeswel/ G4AR
```SW Broadcast Bands . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . Charles Molloy G8BUSMW Broadcast Bands . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . Charles Molloy G8BUSVHF Bands ........... Ron Ham BRS15744
```

KINDLY NOTE-Atomic Time Receiver, August 1977 [R.A.E. 1, September 1977]
For our Constructors
408 SOLO SUPERMIND
At last! Now you don't have to get a friend to set up and check your moves when you want to play the current

```logic game. Use your own random codes on Solo Supermind and then break them-if you can! Alan Willcox
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412 LOW DISTORTION SINE/SQUARE WAVE GENERATOR
This inexpensive AF Oscillator produces either a low distortion sine wave or a $10 \mu \mathrm{~S}$ rise time square wave,

```suitable for all forms of audio amplifier testing.Michael Tooley BA, G8CKT
```

425 GENERAL PURPOSE SW RECEIVER-2
This simple preselector uses a 40673 transistor and is intended for use with the receiver described last month,

```but add a 9 V battery and it can be placed ahead of any SW receiver.Frank Rayer G3OGR
```

426 THE PW "JUBILEE" ELECTRONIC ORGAN-2

```Now that you have got the bits and pieces together for the organ we can begin the constructional work thismonth.
```

```DESIGN YOUR OWN PROJECTS-2
```

Details this month on designing a C

```Toby Bailey and Bob Whitaker
```

AUDIO LEVEL INDICATOR

```Cheap light-emitting diodes replace an expensive meter in this programme level monitor-a must for theaudio enthusiastWilliam Pleass BA, B.Sc, AMIEE
```


## General Interest

## SO YOU WANT TO PASS THE RAE?-2

This month there are a few problems in maths and then the authors get down to Ohm's Law and the use of resistors in electronic circuits $\qquad$ John Thornton Lawrence GW3JGA and Ken McCoy GW8CMY

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| 600mA TOI8 CASE |  |  | 7 AMP TO48 CASE |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 10 | THY600/10 | 60.15 | 50 | THY7A/50 | 60.48 |
| 20 | THY600/20 | $40 \cdot 16$ | 100 | THY7A/100 | 60.51 |
| 30 | THY600/30 | 20.20 | 200 | THY7A/200 | 60.57 |
| 50 | THY600/50 | c0. 22 | 400 | THY7A/400 | 40.62 |
| 100 | THY600/100 | 60.25 | 600 | THY7A/600 | 40.76 |
| 200 | THY600/200 | 60.38 | 800 | THY7A/800 | <0.92 |
|  | THY600/400 | CO. 44 |  |  |  |
|  |  |  | 10 | AMP TO48 | SE |
| 1 AMP TOS CASE |  |  | Volts | No. | Price |
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| Volts 50 | No. | Price | 100 | THY10A/100 | 50.57 |
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| $\begin{aligned} & 600 \\ & 800 \end{aligned}$ | THYIA/600 | 60.45 |  |  |  |
|  | THYIA/800 | 10.58 | ${ }_{\text {Volts }}^{16}$ | AMP TO48 CASE |  |
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## BACK NUMBERS

We are very glad to announce the re-establishment of a PW Back Numbers Service for our readers. In future back numbers dated from June 1977 only will be available from our Post Sales Department for 65p, which includes postage and packing. Cheques and Postal Orders should be made payable to IPC Magazines Ltd.
Send your orders to:- Post Sales Department, IPC Magazines Ltd., Lavington House, Lavington Street, London SE1 OPF.

I SPY!

From the correspondence that we have with our readers it is very evident that one of the principal reasons for the failure of a project to work the first time that it is switched on is the lack of attention to good and proper soldering and the avoidance of solder bridges. On stripboard, particularly that with 0.1 in . matrix, and on some printed circuit boards, the copper rails can be pretty close together and a soldered joint can easily spread until it actually touches an adjacent rail.

When assembling a complicated circuit board it is usual to solder, say, half-a-dozen joints at a time and, in my own case, to inspect them with a watchmaker's magnifying glass. But this is not quite enough. A subsequent joint may lie very close to one made earlier and cause trouble. So it is imperative to check all the board after it is completely finished. This is the point where excitement runs high at the thought of switching on for the first time!

To help you in this checking operation, and indeed at many other points in the course of construction, we are presenting you this month with a magnifying "glass" that could save you from a lot of problems in the future. Keep it by your workplace, preferably in its case to protect it, and you will not regret having bought this issue of Practical Wireless.

Next month's PW will include yet another gift for our readers, in the form of a very comprehensive Component Source Directory, which we believe will prove of inestimable value to the constructor of electronic equipment . . . Indeed, we feel that its appeal may spread much further afield including, as it does, tools and hardware needed in other constructional endeavours. To give you some idea of the amount of effort that has gone into preparing the Directory some 210 copies of an 11-page questionnaire were sent out to a selection of advertisers from PW and other electronic magazines. Some did not bother to reply so the loss of a bit of free advertising is theirs!

Naturally many of the firms shown in the Directory carry a much wider range of goods than it is possible to indicate so do not hesitate to contact a likely supplier for further information. Incidentally, do send a stamped, self-addressed envelope of adequate size for the information you require. With the very high postal charges of today this little gesture is much appreciated by advertisers especially the smaller firms. It can also make quite a difference in the time you may have to wait for a reply!

So don't forget to order your copy of the November issue of PW with its Directory. Keep and treasure the Directory, consult it and use it to your advantage when you are forking out on your hobby! It could save you a lot of time and quite a bit of money in the long run.

Eric Dowdeswell Assistant Editor

## PLEASE NOTE

We do not operate a Technical Query Service except on matters concerning constructional articles published in PW. We do not supply service sheets or information on commercial radios, TV's or electronic equipment.
All queries must be accompanied by a stamped self-addressed envelope otherwise a reply cannot be guaranteed.

## R.A.E. Courses

Mid-Cornwall College of Further Education, Palace Road, St Austell, Cornwall.

Tuesday Evenings 7 to 9 pm commencing 27th September. Enrolment on the 21 st or 22nd September between 5 pm and 7.30 pm . Further information from G4DND. Tel: St Columb 479.

## Boreham Wood College of Further Education, Elstree Way, Boreham Wood, Herts.

Tuesdays and Thursdays 7 to 9pm commencing 20th September. Enrolment on Wednesday or Thursday the 7th or 8th September between 4pm and 6 pm . Further information from G. L. Benbow G3HB. Examination to be sat in December 1977.

## Gosforth Adult Association

 Classes, Gosforth High School, Gosforth, Newcastle-upon-Tyne.Tuesday evenings 7pm to 9 pm with Morse classes on Thursdays at the same times. Further information from D. R. Loveday G3FPE. Tel: Newcastle-upon-Tyne 668439.

Bridgnorth College of Further Education, Stourbridge Road, Bridgnorth, Salop.

Wednesday evenings 7 pm to 8.45 pm

- commencing 14th September. Enrolment on Wednesday or Thursday

7th/8th September between 6.30 and 8.30 pm . Cost for the three term course will be $£ 10$ or $£ 5$ for students under 18. The class tutor will be P. Edwards G3DKJ while the college call sign is G4COB. Further information from R. A. Buckley, Tel: 4431

College of Technology and Design, Feilden Street, Blackburn.

The course will be taken by Harry Leeming G3LLL, and any enquiries should be addressed to The Principal of the College.

North and West Farnborough Further Education Centre, Cove School, St John's Road, Farnborough, Hants.

Thursday evenings at 7.30 pm commencing 22nd September. There will also be a Morse proficiency course beginning on Monday 19th September at 7.30. Further information from J. Brett Principal. Tel: Farnborough 42397.

Knottingley High School, Knottingley, West Yorks.

Tuesday evenings at 7pm. Enrolment on Monday 19th September. Course to be taken by G3HCW.

Newport Amateur Radio Society, Brynglas House, Brynglas Hill, Newport.

Monday evenings 6 pm to 8 pm commencing 19th September. Course fee will be in the region of $£ 7 \cdot 00$, and course tutor will be L. A. Groucott GW3YTJ.


## British Radio Technical Rdvisory Service

The object of this service is to provide a much-needed source to which professionals, transmitting amateurs and short wave listeners and others involved in radio communications may turn for advice. There is a panel of engineers, each a specialist in his field, both in radio communications and electronics in general, and a lifetime of experience to call upon. Questions and requests for advice will be referred to the engineer considered the most experienced in the particular field of the enquiry.

A quote in advance will give the fee for answering questions or requests for advice. This quotation will be offered in two ways, for a short answer or opinion, or for a detailed answer or opinion, based upon full consideration of all aspects of the enquiry and stating the reasons for their opinion.

The only initial outlay to obtain the quotation will be three stamped envelopes, two self-addressed and one left blank to send the enquiry to the appropriate engineer. A quotation for the necessary technical advice or opinion will then be sent. Upon receipt of the remittance the necessary technical advice or opinion will be sent together with a suggested source and, where possible, the cost of any equipment to comply with the needs.
G2DYM Aerials and Projects, Whiteball, Wellington, Somerset.

## Wessex report

The Wessex AR Group had the good fortune to have two interesting lecturers recently. Ken Alford G2DX, holding an early TH500 valve, talked about the early days of Wireless and F.J.H. "Dud" Charman G6CJ, extreme right, gave his well-known lecture on Aerials. Between them is Frank HicksArnold G6MB and Roy Scott G2 CZH is at the left, President and Vicechairman of the Group respectively.
Hon. Sec. G. Coles G4EMN, 6 St. Anthony's Road, Bournemouth.

## 0000

The popular game of cunning and logic called "Mastermind" is intended for two players, named respectively, the "Codemaker" and "Codebreaker". Solo Supermind carries out all the necessary operations allowing the game to be played by one person. In Mastermind the Codemaker begins the game by placing four coloured pegs-the Code-in any order but hidden from the Codebreaker. There are six colours to choose from, giving 1,296 permutations. The object of the game is that the Codebreaker should duplicate the exact colours and positions of the secret Code.
For those not familiar with Mastermind the sequence is as follows:-the Code set by the Codemaker at the top of the board, shielded off, might be yellow, red and two blues. The Codebreaker begins his play by placing four coloured Code Pegs in Row 1-a pure guess at this stage. Suppose he has tried green, blue, yellow and red. The Codemaker must now give information by placing black and white "Key Pegs" in the Key Peg holes alongside the Code Pegs. A black Key Peg is placed in any of the Key Peg holes for every Code Peg which is of the same colour and in the same position as one of the Code Pegs in the hidden Code. A white Key Peg is awarded for each Code Peg of the right colour but in the wrong position. In this case three white Key Pegs are awarded by the Codemaker for the blue, yellow and red Pegs which are the correct colour but in the wrong positions.
On the basis of this information the Codebreaker tries another four Code Pegs in Row 2, hopefully bringing him nearer the solution. The three white Key Pegs give no indication as to which three Code Pegs are the correct colour, and so the Codebreaker in Row 2 tries retaining yellow, green and blue, and replaces the red with an orange. In the process he

## components

## Resistors

$\left.\begin{array}{ll}\text { R1 to R8 } \\ \text { R9 to } 12 & 2 \cdot 2 \mathrm{k} \Omega 2 \% \\ \uparrow \cdot 5 \mathrm{M} \Omega\end{array}\right\} \frac{1}{4}$ or $\frac{1}{3} \mathrm{~W}$

## Semiconductors

Tria to Tr4b BCY 71 (8 off)
Tr5a to Tr8b BC 109 (8 off)
D1 to D8 general purpose silicon e.g. 1 N4t48
Miscellaneous
Sf to S8 2-pole 6-way midget wafer. Doram code 327-254 S9 push-to-make switch
Case $\quad \begin{aligned} & \text { plastic } 216 \mathrm{~mm} \\ & \text { code } 509-608\end{aligned} \times 137 \mathrm{~mm}\left(8 \frac{1}{2}^{\prime \prime} \times 5^{*}\right)$ Doram
Meters, 1mA FSD (2 off). Control knobs, 4 plain, 4 pointer. Battery clip. 9V Battery PP3. 6 B.A. bolts, nuts and spacers (each 4 off). Veroboard $0.1^{\prime \prime}$ matrix- 79 mm $\times 28 \mathrm{~mm}\left(3 \frac{1^{\prime \prime}}{} \times 1 \frac{1{ }^{\prime \prime}}{4}\right)$

has also shifted them about in order to try to achieve the correct position. The Codemaker replies by awarding two black Key Pegs, one each for yellow and blue, these being the correct colour in the correct position; but two Key Peg holes are now left vacant because orange and green do not appear at all in the Code. From this information the Codebreaker can deduce that red appears in the hidden Code and orange does not.

And so the game progresses, with the Codebreaker continually referring back to the results of previous tries in order to decide the best arrangement for the next move. When, in the illustrated case, the Codebreaker has reached Row 4 he can, by referring back, say with certainty what the hidden Code is, which he does in Row 5.

These logical thought processes can take a considerable time, and the Codemaker is apt to become impatient, all the more so as his role is rather a boring one! For this reason, as Mastermind players will confirm, it is often difficult to find an opponent willing to take on the role of Codemaker. Solo Supermind is designed to replace the Codemaker entirely, so allowing the game to be played by one person and at his own pace.

## Using Solo Supermind

Switches S1 to S4 have their stops removed to allow them to rotate continuously, and at the commencement of the game these are turned at random to set the hidden Code, then left alone throughout the game. These four Codemaker switches are operated by control knobs which are unmarked, and in practice they have proved to be a simple and effective way of obtaining a random code.
The game then proceeds in the normal way with the first guess being entered in Row 1 on the Code-
maker board. In order to find out how many and which Key Pegs are due for this attempt the calibrated Codebreaker switches are turned to indicate the same colours as the Code Pegs, the left hand switch corresponding to the left hand Code Peg, and so on. When the push switch is depressed the number of Key Pegs to be awarded is shown on the two meters, one showing the black and the other the white Pegs. The indicated Key Pegs are placed in position on the Mastermind board and the game continues, each attempt being duplicated on Solo Supermind to obtain the number of black and white Pegs due at each stage. The game is over when the black meter indicates "four", and of course the aim is to reach this with the least number of attempts.
Current is drawn by the circuit only when the push button is operated, and is then only 1 mA max. so the battery, once installed, can virtually be forgotten. Keen Mastermind players will find their enjoyment of the game much enhanced with Solo Supermind for, freed from the disturbing presence of another person waiting, patiently or impatiently, for a move to be made, concentration and logical attack are much easier to maintain.

## Circuit description

The key to the simplicity of the circuit, Fig. 1other designs have used 40 or more IC's-is the con-
stant current technique employed in the stages Trl to Tr8. Each of these stages will allow only one "unit" of current to flow, and these are counted by the meters. S1 to S4 are the "hidden" Codemaker switches, and S5 to S8 are the Codebreaker switches which duplicate the Code Pegs on the Mastermind board.

Maximum current through each stage is limited to $250 \mu \mathrm{~A}$, since, at this point, 0.5 V is developed across each emitter resistor, turning on the complementary transistor and therefore removing bias. The advantage gained from this is that the circuit is not critical of transistor gains, and the meter pointer shows no variation of position for different switch combinations deserving the same value. Hence, the scale can be marked in points rather than areas, and remains compatible with $\operatorname{lmA}$ FSD.

The bias resistors R9 to R12 are high in value so that the current via them to the white meter in the open collector situation, and to the black meter via D5 to D8 when it applies, is insignificant.

## Switching sequence

The position of Sl corresponds to the colour of the left hand hidden Code Peg. Now if S5, a Codebreaker switch, in an attempt to duplicate the Code, is switched to the same position as S1 this would count as a black Peg, corresponding to the correct colour in the correct (left hand) position. One unit


Fig. 1: The complete circuit diagram, showing the extensive use of constant current stages.
( ${ }^{1} 4 \mathrm{~mA}$ ) of current will flow through S1a and S5a to the black meter which, if the sensitivity is 1 mA FSD will register a quarter FSD corresponding to one black Peg. Note that no current flows through D1 because the voltage developed across the meter is too small to allow it to conduct. So D1 serves to isolate Trl's current and the path to M1 via Sla from the other switch banks.

Similarly, if any of the other switch pairs S2/S6, S3/S7 and S4/S8 are in the same position, current will flow through their ' $a$ ' sections to be counted by the black meter M1. Just as the ' $a$ ' sections are concerned solely with the allocation of black points, so the ' $b$ ' sections are concerned with the white. Say the Codebreaker switch S 5 is positioned corresponding, not to S1 but to either S2, S3 or S4, then this would signify correct colour but wrong position and deserve a white Peg. In this case current is only available via the ' $b$ ' sections, which have all their similar positions (colours) linked together, so that the unit of current, whether supplied by $\operatorname{Tr} 2, \operatorname{Tr} 3$, or Tr4 is available on all the ' $b$ ' sections to pass, in this case via Tr5, to the 'white' meter M2.

## Components

The meters used on the prototype were inexpensive units of the type commonly used as audio level indicators, with a sensitivity of approximately $250 \mu \mathrm{~A}$ FSD; but almost any meter will do the job provided it is small enough to fit into the case and has sensitivity up to 1 mA FSD. The circuit supplies 1 mA for FSD,
and so if more sensitive meters are used some of this current must be shunted around the meter. There is not much room to spare in a case of the kind used in the prototype, so if more conventional larger panel meters are used the unit will have to be housed in a roomier case.

## Construction

The case specified is supplied with a metal front panel, although a plastic one may be preferred in that it tends to ease the hole cutting problem, reduces the risk of shorts, and provides a neater, more durable finish. The front panel size is $156 \times 91 \mathrm{~mm}$, and if the meters chosen are available a check should be made before the case is purchased that this area is sufficient to accommodate them and the switches. The area required for the switch bank is $125 \times 70 \mathrm{~mm}$.

The colour indicators for the Codebreaker switches were made on the prototype by using Code Pegs from a Mini-Mastermind. These were pushed into $3 / 32^{\prime \prime}$ diameter holes from the rear of the panel until just protruding, and then snipped off. The recommended switches have an adjustable stop which is removed for the Codemaker switches S1 to S4. Also there is a fixed stop consisting of an indentation in the switch body which must be removed by filing away or cutting to allow these four switches to rotate continuously. In the case of the Codebreaker switches the fixed and adjustable stops are retained to give six positions.

The switch wiring is quite straightforward as long as care is taken to position the switches exactly as


Fig. 2: The switch wiring, meter, and transistor connections.


## Underside view of com-

 pleted unilt showing positions of codemaker (upper) and codebreaker (lower) swiches retative to meters.shown in Fig. 2. Examination will show that it is as though the switch pairs were connected directly back to back, and so they have an opposite sense of rotation. That is, the sequence of colours which is clockwise for the Codebreaker switches is anticlockwise for the Codemaker switches. The second stage of the switch wiring is to connect together corresponding positions of the ' $b$ ' sections of the Codebreaker switches. The wiring of the switch poles and the interconnections are also shown in Fig. 2.

The remainder of the components are assembled on a small piece of veroboard as shown in Fig. 3. When mounting the components ensure that their height measured from the surface of the board does not exceed ${ }^{1}{ }^{\prime \prime}$, otherwise it will be difficult to fit into the
space provided. The battery is held in place by a clip, whilst the board is retained in the case by 6 BA bolts, nuts and spacers.

## Playing the game

When starting a game the Codemaker switches, rather than just being turned at random, can be set by another person, to perhaps what he feels to be a difficult combination. If Solo-Supermind is played without a Mastermind board, using pen and paper to record the moves, it is more convenient to use a number code than a colour one. The Codebreaker switches would then be marked with the numbers 1 to 6 instead of the colours shown.

Fig. 3: Top view of veroboard showing overall component layout. Note matrix alphabetical and numerical references, and breaks in print.

Note: Break print at $H, O, V$ at numbers $2,3,4,7,8,9$,


## LOW DISTORTION sine/squanewave GENERATOR

## MICHAEL TOOLEY BA/G8CKT

A variable frequency and variable voltage signal source is a most useful addition to the test equipment in an experimenter's workshop. This signal generator, suitable for general audio frequency work, provides both sine and square wave output covering the frequency range 15 Hz to 25 kHz . The output voltage is continuously variable from 0 to 3 V peak-to-peak.

The attenuator output level calibration is adequate for testing high and medium impedance circuits. However, a voltage follower is also described for addition when the signal generator is to be used with low impedance circuits.

## Circuit description

The circuit uses a Wien bridge network to determine the operating frequency. The basic circuit of a Wien bridge is shown in Fig. 1. The frequency determining components, C and R , are connected in series in one arm of the bridge and in parallel in the adjacent arm. If a sine wave alternating voltage is applied to the network at terminals A-B, the voltage appearing between terminals C-D will be out of phase by an amount determined by the values of C and R . If the frequency of the voltage is varied (whilst the values of $C$ and $R$ remain fixed) the resulting phase


Hge $\boldsymbol{H}$ (left The basic arrangement of a Wheatstone bridge incorporates: a bal anced reactive ef oment m one:arm 10 form a Wen brioge: hefwork

19e 2 (6) Wein ob mb nelyop as usicis ols iofrem ofthisosctionor

shift produced by the network will also vary. At a certain frequency the network will produce zero phase shift and the voltage between $C-D$ will be in phase with the voltage A-B. This frequency is given by the relationship:
$f=\frac{1}{2 \pi C R}=\frac{0 \cdot 159}{C R} \mathrm{~Hz}$. (with C in farads and R in ohms).
The amplitude of the voltage appearing between $C-D$ is determined by the resistive arms of the bridge, R2 and R3. The values of R2 and R3 may thus be adjusted to give a desired level of output.
The simplified circuit of the sine wave generator is shown in Fig. 2. An amplifier is used in conjunction with the Wien bridge network. The input of the amplfier is taken from C-D and the output is taken to A-B. Provided that the amplifier gives sufficient gain to overcome the loss in the bridge, and that the amplifier exhibits an overall phase shift of $0^{\circ}$ or $360^{\circ}$, the feedback will be positive at the bridge frequency and oscillation will result.
The resistance $\mathbf{R}$ is varied continuously by the use of a dual-gang potentiometer, VR1A/B. The values of capacitance, C , may be conveniently switched in decades, $\mathrm{C} 1, \mathrm{C} 2, \mathrm{C} 3$ and $\mathrm{C} 4, \mathrm{C} 5, \mathrm{C} 6$ to give three frequency ranges while preserving the basic scale calibration. Two equal fixed resistors (R2 and R3 in Fig. 2) restrict the frequency coverage obtained in any one range to a sensible amount while ensuring a reasonable overlap between the ranges.
Manual adjustment of the amplitude of oscillation is provided by the variable resistor, RV. If the amplitude of oscillation is allowed to be too large, appreciable distortion will result. Automatic amplitude control is also provded by the use of a thermistor,

## specifications




Fig. 3: Circuit diagram of the AF Signal Generator. The voltage follower circuitry surrounding IC1 may not be required (see text 'Construction').

## components



RTH, in the feedback path. The thermistor keeps the amplitude of the output within close limits over the entire operating frequency range of the oscillator. It should be noted that, in the complete circuit of Fig. 3, only AC negative feedback is applied through the thermistor. The DC conditions are not stabilised by means of the thermistor since this can give rise to objectionable amplitude 'bounce' consequent on an adjustment of operating frequency.

Transistors $\operatorname{Tr} 1, \operatorname{Tr} 2$ and $\operatorname{Tr} 3$ form a three-stage amplifier with direct coupling between the stages. Switch S 2 selects either sine or square wave output. In the square wave position an extra amplifier stage, $\operatorname{Tr} 4$, is introduced. This amplifier stage is substantially overdriven and effectively clips the sine wave. The clipping action is aided by D1 which also preserves the symmetry of the square wave. The output level on square and sine wave may be adjusted by means of VR4 and VR5, respectively. The sine or square wave output from S2b is fed to a simple switched resistive


Fig. 4 (above): Power supply. C11 is mounted external to the PCB. Fig. 5 (below): The optional voltage follower circuitry. Simply leave these components off the PCB if not required.


attenuator. The lower resistor of the potential divider network is made variable, thus providing a continuous adjustment of the output voltage level.
The power supply uses a conventional full-wave bridge rectifier arrangement. The current consumption of the signal generator is small and remains constant thus no provision for DC stabilisation is incorporated.

## Construction

Although the prototype was constructed on veroboard, the design was considered eminently suitable for conversion to a printed circuit board. All three sections of the circuit (see oscillator, Fig. 3; power supply, Fig. 4 and voltage follower, Fig. 5) are incorporated on a single board to facilitate easy construction. Some readers may not require the voltage follower circuitry in their application. In this case, it is recommended that the associated components are simply left off the board and the output taken directly from the slider of VR2.
Fig. 6 shows details of the copper side of the PCB while Fig. 7 gives component identification and position. The ready-etched board may be obtained from the PW Readers PCB Service for those people who do not wish to make their own.

It is suggested that multiway ribbon cable is used to connect the front panel controls to the circuit board. This method of wiring allows much neater construction and minimises the risk of wiring errors at the same time.

## Adjustments and calibration

Set VR1 to mid-position and S1 to the middle range, a frequency of approximately 1 kHz . If an oscilloscope is available, adjust VR3 for a continuous sine wave oscillation with no noticeable distortion. Otherwise, connect the signal generator to an audio amplifier and adjust the volume control for a comfortable level of output. Set VR3 to maximum resistance and slowly rotate the control until oscillation begins. Note this position carefully. Continue rotating VR3 in the same direction until the oscillation just ceases and note this new position. Set VR3 mid-way between the two positions. Oscillation should then be continuous and undistorted. As a rough guide, the VR3 slider should be at about mid-travel, providing a resistance of between $220 \Omega$ and $300 \Omega$.

Adjust VR2 for maximum output and switch S3 to the ' 3 V ' range. If an oscilloscope is available, adjust VR5 and VR4 for exactly 3V peak-to-peak output on sine wave and square wave, respectively. Alternatively, a high impedance AC voltmeter may be used. If the instrument is calibrated in RMS rather than peak-to-peak units, the corresponding outputs are approximately 1.1 V sine wave and 1.5 V square wave.

The output frequency may be checked either by using a digital frequency meter or by using an oscilloscope with a calibrated timebase. Another method, which is often described in text books on electrical measurements, involves the use of a signal at a known frequency to display Lissajous figures on the screen of an oscilloscope. It should only be necessary to calibrate the signal generator on the middle frequency range, 150 Hz to $2 \cdot 5 \mathrm{kHz}$, the other ranges can make use of the same scale markings multiplied or divided by ten accordingly.

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The passing of the Radio Amateurs' Examination, set by the City and Guilds, requires a certain level of theoretical technical knowledge. Whether one considers that this level is too high or too low is beside the point. The course that follows is intended, with the help of certain external aids, to prepare the reader to pass the examination. It will not teach him all about electronics!

## Any questions?

Before we leave calculations and move into the next section, are there any queries? Yes! When finding the square root of indices, by halving the index number, what happens if the index number is not divisible by two? OK, in this case it is necessary to rearrange the values to obtain an index number which can be halved. For example, suppose we have $\sqrt{8 \cdot 1 \times 10^{5}}$ rearrange by reducing $10^{5}$ by a factor of 10 , to $10^{4}$, and multiplying $8 \cdot 1$ by this factor, we get $\sqrt{ } 8.1 \times 10 \times 10^{4}=\sqrt{81 \times 10^{4}}$. Square root the $81=$ $9 \times \sqrt{10^{4}}$. Halve the index number $=9 \times 10^{2}$ bring to normal notation $=900$.

If the index number were negative, for example $\sqrt{8 \cdot 1 \times 10^{-5}}$ then rearrange by reducing $10^{-5}$ by a factor of 10 , to $10^{-6}$, and again multiplying $8 \cdot 1$ by this factor

$$
\begin{aligned}
& =\sqrt{8.1 \times 10 \times 10^{-6}} \\
& =\sqrt{81 \times 10^{-6}} \\
& =9 \times \sqrt{10^{-6}} \\
& =9 \times 10^{-3} \\
& =0.009
\end{aligned}
$$

## Attention please!

The most important single factor in preparing for the RAE is to get plenty of written practice and this includes working through calculations, drawing diagrams and writing explanations. It is so important that we will say it again! Plenty of practice is the key to success.

Got the message? Good! Then how about having a shot at answering the following by writing them out, stage by stage. The answers are given at the end of this part.

1. Express 0.001 microfarads ( $\mu \mathrm{F}$ ) in picofarads ( p ) .
2. Express $3 \cdot 6$ millihenrys ( mH ) in microhenrys $(\mu \mathrm{H})$.
3. Express $33 \times 10^{2}$ ohms ( $\Omega$ ) in kilohms ( $\mathrm{k} \Omega$ ).
4. Express 1296 megahertz ( MHz ) in gigahertz ( GHz ).
5. Solve $\sqrt{6.4 \times 10^{7}}$.

## Current

Before discussing resistance and resistors, let us remind ourselves about current flow. You know that all materials are made up of atoms, each having a

. . . still having trouble with square roots?
positively charged nucleus, and electrons with a balancing negative charge orbiting it. Current flow through a material is based on the movement of this negative charge from one atom to another and this depends on how loosely or tightly the electrons in each atom are bound to the nucleus and how much external attraction (applied voltage) there is to move them.

Conventional Current and Electron flow
Historically it had always been assumed that an electric current flowed from the positive terminal of the supply, through the external circuit and back to the negative terminal. However, with further knowledge of the structure of the atom it became obvious that current was due to the movement of a negative charge towards the positive (as in the radio valve). This apparent contradiction has caused much discussion over the years, but the accepted view and the one we will use is that, as stated previously, conventional current flows from the positive to the negative irrespective of the actual method or mechanism of the flow.

You may have noticed that it is also common practice to draw diagrams with the positive supply line at the top of the page and the zero or negative at the bottom. This convention makes it much easier to visualise the flow of current always coming vertically down the page through the various parts of the circuit. We will be discussing this in more detail later in the series.

## Water Flow Analogy

The usual analogy to electrical flow in a circuit is water flow through a pipe. The head of water or pressure represents the voltage, the flow of water through the pipe represents the current and the resistance to flow, caused by the smallness of the pipe, is equivalent to electrical resistance. The analogy also holds good in that, for example, when a tap, connected to a full hosepipe, is opened, water flows immediately out of the other end of the pipe and yet it is some time before a particular drop of water leaving the tap emerges from the far end. Similarly, an electric current entering a wire appears at the far end almost instantaneously but the actual 'bits' of charge forming the current take an appreciable time to make their way from atom to atom along the whole length of the wire.

## Conductors and insulators

Those materials having atoms with loosely bound electrons, which can move easily from atom to atom, have a low resistance to current flow and are known as conductors. All metals are good conductors. Those materials having atoms with tightly bound electrons, which only move when under great electrical stress, have a very high resistance to current flow and are known as insulators. For example, an electric cable has a copper core to allow a free flow of current along its length and a plastic or rubber sleeve to insulate the core and to prevent current from leaking away to adjacent wires or to you, if you happen to be holding it!

Here is a list of typical conductors and insulators:-

| Conductors | Insulators |
| :--- | :--- |
| Silver | Mica |
| Copper | Quartz |
| Aluminium | Glass |
| Brass | Ceramics |
| Iron | Plastics |
| Mercury | Rubber |
| Carbon | Oil |
| Some liquids | Air |


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## Ohm's Law and all that

In radio and electronic circuits, besides needing good conductors and good insulators, we need a range of resistors which have a known, marked, value of resistance that does not change with temperature or time. The actual value of resistance is chosen to suit the requirement of the circuit in which it is to be used and this is calculated using Ohm's Law.
You may remember that Ohm's Law states that for a particular resistor, the ratio of the voltage across that resistor, to the current flowing through it, is constant. Thus, if we increase the voltage across the resistor, the current flowing through it will also increase, but the ratio between the two will remain constant. This ratio is known as the resistance and is stated in ohms ( $\Omega$ ), thus:-

## $\frac{\text { VOLTAGE }}{\text { CURRENT }}=$ Constant $=$ RESISTANCE, ohms ( $\Omega$ ).

At this stage, a small memory aid is appropriate,


By covering the unknown quantity, the appropriate formula is shown. From this you can write down the relationships between the three quantities,
CURRENT (I) VOLTAGE (V) and RESISTANCE (R).
So you have

$$
\begin{array}{ll}
\mathrm{I}=\frac{\mathrm{V}}{\mathrm{R}} & \text { Current }=\frac{\text { Voltage }}{\text { Resistance }} \\
\mathrm{R}=\frac{\mathrm{V}}{\mathrm{I}} & \text { Resistance }=\frac{\text { Voltage }}{\text { Current }} \\
\mathrm{V}=\mathrm{I} \times \mathrm{R} & \text { Voltage }=\text { Current } \times \text { Resistance } .
\end{array}
$$

Another quantity occurs here, POWER. This is the energy dissipated as heat when a voltage is applied across a resistance and causes a current to flow through it. The power is the product of the voltage and the current,

Power $(W)=$ Voltage $\times$ Current $=V \times I$.
Looking at this, you will no doubt spot the fact you can write another set of relationships and bring in another memory aid,


Code word WIV

It is then possible to combine them and produce a number of relationships or equations relating to Power, Voltage, Current and Resistance.

$$
\begin{aligned}
& \text { Power (watts) } W=V \times I \\
& \text { but } I=\frac{V}{R} \\
& \text { so that } W=\frac{V \times V}{R}=\frac{V^{2}}{R}
\end{aligned}
$$

$$
\begin{aligned}
& \text { also } V=I R \\
& \text { so that } W=I R \times I=I^{2} R
\end{aligned}
$$

summing up, you have $\mathrm{W}=\mathrm{VI}=\mathrm{I}^{2} \mathrm{R}=\frac{\mathrm{V}^{2}}{\mathrm{R}}$
The composite diagram or formula wheel given below, is rather too complex to be a memory aid but it is, nevertheless, very useful to refer to.


Here is an example of the memory aids in use.


Given the voltage, and the resistance, calculate the current.

$$
\mathrm{I}=\frac{\mathrm{V}}{\mathrm{R}}=\frac{12}{6}=2 \text { amperes. }
$$

Given the voltage and the current, calculate the power.

$$
\mathrm{W}=\mathrm{V} \times \mathrm{I}=12 \times 2=24 \text { watts. }
$$

For your own reassurance, you may care to try out other formulæ with the values given.

## Resistors in Series and Parallel

It will do no harm at this stage to have a brief reminder of how resistors behave in series and in parallel combinations.


Total Resistance $\left(\mathrm{R}_{\boldsymbol{Y}}\right)=\mathrm{R} \mathbf{1}+\mathrm{R} 2+\mathrm{R} 3$.


$$
\frac{1}{R_{1}}=\frac{1}{R 1}+\frac{1}{R 2}+\frac{1}{R 3}
$$

or if there are only two resistors,
$\mathrm{R}_{\mathrm{T}}=\frac{\mathrm{R} 1 \times \mathrm{R} 2}{\mathrm{R} 1+\mathrm{R} 2}$

Where both combinations occur it is known as the series-parallel case.


If you can remember the two memory aids given previously, you have all that is necessary to sew up the DC conditions in a circuit.

## Potential and Current Dividers

There are two more useful cases which are variations of the basic DC circuit that are worth remembering, namely, the Potential Divider and the Current Divider. The potential divider is useful when setting up bias circuits for the base of a transistor, and the current divider when it is required to have currents flowing in two parts of a circuit.


Here are three more simple problems for you to do. Please write everything out in full. Answers at the end.
6. Find the total resistance of this series-parallel arrangement.


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1. Wirewound Resistor 50W. 2. Carbon film resistor 30W. 3. Wirewound resistor 10W. 4. Wirewound resistor 5W 5. Wirewound resistor $2 W .6,7,8$ and 9. Carbon film resistors, $2 W, 1 W, \frac{1}{2} W$ and $\frac{1}{4} W .10,11$
and 12. Potentiometers, $3 W$ wirewound, $1 W$ carbon and pre-set carbon.



Resistor colour code table. Decode value of resistor as shown above. Note the third figure, the multiplier, is the number of noughts after the first two significant figures. Tolerance is usually indicated by an additional band, gold for $5 \%$ and silver for $10 \%$.
wound resistor may be marked $100 \Omega \pm 5 \%$, which tells you that the actual value will be within 95 and $105 \Omega$.

A new method of stating the resistance is given in the BS1852 resistance code, as shown below:-
$0 \cdot 56 \Omega$ would be R56
$1 \cdot 0 \Omega$ would be 1R0
$5 \cdot 6 \Omega$ would be 5 R 6
$56 \Omega$ would be $56 R$
$100 \Omega$ would be 100 R
$1 \mathrm{k} \Omega$ would be 1 K 0
$10 \mathrm{k} \Omega$ would be 10 K
$10 \mathrm{M} \Omega$ would be 10 M

After the value, a further letter is added to indicate the tolerance, $\mathrm{F}= \pm 1 \% \mathrm{G}= \pm 2 \% \mathrm{~J}= \pm 5 \% \mathrm{~K}= \pm 10 \%$ $\mathrm{M}= \pm 20 \%$. For example:-

$$
\begin{array}{ll}
\mathrm{R} 56 \mathrm{M}=0 \cdot 56 \Omega \pm 20 \% & 68 \mathrm{KK}=68 \mathrm{k} \Omega \pm 10 \% \\
390 \mathrm{RJ}=390 \Omega \pm 5 \% & 1 \mathrm{~K} 2 \mathrm{~F}=1 \cdot 2 \mathrm{k} \Omega_{ \pm} \%
\end{array}
$$

For the purposes of the R.A.E, either method would be equally acceptable, but it would be a good idea to stick to the method with which you are most familiar and use this exclusively.

All other resistors have their value and tolerance marked by coloured bands printed on the body of the resistor. The resistor colour code is given on the Practical Wireless INFO-CARD (March '77).*

Variable resistors are usually made in the form of a potentiometer with connections to both ends of the resistance element and to the sliding connection. Carbon is used as the resistance element in potentiometers up to about 1 watt rating and higher power versions are almost invariably wire-wound, 3 watts being a power size.

* Practical Wireless INFO-CARD (March '77) is avail. able free of charge. If you would like one please send a stamped, self-addressed envelope, at least $9 \times 7 \mathrm{in}$, to INFO-CARD, c/o Practical Wireless, Fleetway House, Farringdon Street, London EC4A 4AD. Do not enclose any correspondence.

. . . . your friendly local amateur.


## Answers to problems

1. 1000 pF
2. $3600 \mu \mathrm{H}$
3. $3 \cdot 3 \mathrm{kilohm}(\mathrm{k} \Omega$ )
4. $1 \cdot 296 \mathrm{GHz}$
5. 8000
6. $\mathrm{R}_{\mathrm{T}}=\mathrm{R} 1+\frac{\mathrm{R} 2 \times \mathrm{R} 3}{\mathrm{R} 2+\mathrm{R} 3}=8+\frac{3 \times 6}{3+6}=8+\frac{18}{9}=8+2=10 \Omega$
7. $I=\frac{W}{V}=\frac{30}{12}=2 \cdot 5 \mathrm{~A}$
8. $I=\sqrt{\frac{W}{R}}=\sqrt{\frac{40}{10}}=\sqrt{4}=2 \mathrm{~A}$
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| $\mathbf{S S . 3 2 4}$ | $24 \mathrm{~V} / 1 \mathrm{~A}$ | $\mathbf{£ 7 . 6 5}$ |
| $\mathbf{S S . 3 3 4}$ | $\mathbf{3 4 V} / 2 \mathrm{~A}$ | $\mathbf{£ 8} \cdot \mathbf{7 5}$ |
| $\mathbf{S S . 3 4 5}$ | $\mathbf{4 5 V} / 2 \mathrm{~A}$ | $£ 10.75$ |
| $\mathbf{S S . 3 5 0}$ | $\mathbf{5 0 V} / 2 \mathrm{~A}$ | $\mathbf{£ 1 1 . 7 5}$ |
| $\mathbf{S S . 3 6 0}$ | $\mathbf{6 0 V} / 2 \mathrm{~A}$ | $\mathbf{£ 1 2 . 7 5}$ |
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## Preselector

Though this preselector is particularly intended for use with the $1 \cdot 7-15 \mathrm{MHz}$ receiver described last month, it can be employed with other receivers, and can if necessary be adjusted to cover approximately 1.5 MHz to 16 MHz . It provides an increase in sensitivity and a substantial reduction in second channel interference.

Potentiometer VR1 in Fig. 8 is an input attenuator which is essential to avoid overloading of the receiver with strong signals. It also allows the receiver input to be kept down, when necessary, for satisfactory reception of SSB and CW signals. The 2 -pole switch S1a/b selects L1 or L2, and these are adjusted to provide a band coverage matching that of the receiver described. As this stage is individually tuned
by VCl, no ganging problem arises. Output from the drain of the 40673 goes to the main receiver via the input switch which selects the appropriate receiver aerial coil primary. These windings are supplied from
continued on page 433




Most of the construction takes place on a single printed circuit board, details of which are shown in Fig. 1. Because of its size, it cannot be shown full size in the magazine. Full size prints for PCB production purposes can be supplied from the editorial address at the front of the magazine, price 40 p . Large (at least $13^{\prime \prime} \times 8^{\prime \prime}$ ) stamped and addressed envelopes must accompany the cheque/money order. As usual, readymade boards will be available from the PW Readers' PCB Service.

## Keyboards

An early decision must be taken regarding choice of calculator key, or conventional keyboard. If you intend to make the former, start by cutting out the corner of the printed board along the guide lines. This
is to allow room for the mains transformer in the final stages of assembly. Continuing with the same version, refer to the PCB overlay Fig. 2 and insert the key switches. Make sure that they are pushed fully home and twist them slightly to get them into line. A smear of Evostick will hold the switches in place while soldering.

Insert and solder all the board pins-their positions are highlighted on Fig. 2. It is important that these are put in before other components are added to prevent the possibility of damage during insertion. For the conventional keyboard version, insert pins into the holes marked within the key switch areas of Fig. 2. These pins are used to take flying leads to the contacts under the keyboard.

The heading photograph shows a conventional keyboard verslon of the Jubilee organ. The PCB is shown in the lower picture. Note the replacement of calculator keys with PCB pins.

Fig. 1: The PCB shown copper side. This drawing has been reproduced at under haif the original size to fit in with the page format of the magazine. The actual PCB size is $50 \cdot 8 \times 24 \cdot 1 \mathrm{~cm}$. Full size prints forial address at the front of the magazine, price 40 . Be sure to include a large (at least $13^{\prime \prime} \times 8^{\prime \prime}$ ) stamped and from the editorial address at the front of the magazine, price 40p. Be surel
addressed envelope with the chequelmoney order.

The board area incorporating the logo at the top left hand side is intended to be cut away to house the mains transformer in the calculator keyboard version where space is at a premium. In a creative moment we envild

## Links

Next solder in all the top side links. It is easiest to use 22SWG tinned copper wire; make sure that there are no kinks, to minimise the possibility of shorts between adjacent links. To straighten, stretch the wire slightly before cutting to length. Fig. 2 highlights details of all links.

The circuits are described at each stage of construction. This should help in a logical approach to stage-by-stage testing. This will enable constructors to test the circuits as they are completed.

## Power supply

The power supply schematic is shown in Fig. 3. Assembly details are shown in Fig. 4. Three power rails plus ground are required; $+12 \mathrm{~V},-12 \mathrm{~V}$ which


are regulated and +15 V unregulated. These are obtained from a double-wound mains transformer (mounted off the board) and miniature DIL bridge rectifiers. When inserting these; check that they are the right way round. Likewise, check the polarity of C44, 45 and 46 . Note that 93 is a $1_{2} \mathrm{~W}$ resistor. The regulators are mounted off the board on the back panel or on a separate heat sink; connections are made via board pins. Note that a mica insulator is required under IC22. The four 470 nF capacitors (C47 to C50) are there to prevent parasitic oscillations in the regulators.

Cross check component insertions against sectional component lists.
Test the supply before progressing further. This requires flying leads to the transformer and regulators (at this stage they need not be on a heat sink). Connect the regulator pins ' 3 ' back to the pin labelled 'Chassis Tag' adjacent to B1. Refer to Fig. 4.

Flying leads should also be connected between the regulators and pins labelled +12 V and -12 V immediately above C49 and C50 and to the right of IC2; also wire between all other connections marked 'chassis tag' in Fig. 4. Connect the mains to T1 and check voltages in respect of Fig. 3.

## Power amplifier

Assemble the power amplifier stage shown in Fig. 5. This uses the LM380 which, in this application, will deliver about 1W into an $8 \Omega$ loudspeaker. Resistors R87, 88 and 89 together with R90 mix the signals from the Drums, Melody and Accompaniment circuits. This is all done at low impedance to avoid undue pick-up of the chorus from the tone generators. It also provides an output via C41 for an external power amplifier. The signal level is approximately 150 mV -ideal for feeding the crystal cartridge or phono input of a


domestic hi-fi system. R91 and 92 attenuate the signal slightly prior to feeding the power amplifier IC20. Ensure that C43 is inserted with the correct polarity sense.
A check may be made to ensure that the power stage is working. Re-connect the flying leads for the power supply and add a couple of leads for the loudspeaker (see Fig. 4). Touching the top end of R92 should produce 50 Hz hum pick-up in the loudspeaker. Apart from this there should be no other noise from the power stage. A correctly functioning power amplifier will allow checks on other signals.

## Generators and dividers

Fig. 6 shows the Master Oscillator, Vibrato and Divider circuits. Trl and $\operatorname{Tr} 2$ form an emitter-coupled multivibrator the frequency of which is controlled by C1 and VR1 in association with R3 and R4. VR1 is used to provide fine tuning and should initially be set in the middle of its track. Its frequency can be shifted slightly by varying the voltage at the emitter of Tr2.


Fig. 5: Power amplifier schematic.

Fig. 4 : Blown up sector of the PCB overlay showing the power supply, audio amplifier and tone generator sections.
A low frequency sinusoidal signal fed to this point provides vibrato. ICl is connected as a low frequency phase shift oscillator to give vibrato. Its frequency is controlled by C3, 4 and 5 in association with R6, 8 and 9 . If the value of VR2 is too low, oscillation will not occur; if too high the vibrato output will distort. VR3 controls vibrato depth.
When assembling the PCB, the use of sockets for all the ICs except IC20 is strongly recommended. Temporarily connect VR3 to the three pins to the right of IC1-their separation allows the potentiometer to fasten directly to the pins. If you own an oscilloscope, check that the oscillator and vibrato stages are working by monitoring the collector of Tr2. Initially set VR3 to the OV end of its track to produce a well defined square wave (frequency 250 kHz ) adjustable over a small range by VR1. Centre this frequency on 250 kHz . Turning VR3 up to maximum should cause no change to the signal; slowly increase the value of VR2 to start the vibrato oscillator. At a certain setting, the signal at VR2 should change frequency at a regular rate.

Switch the power off and re-apply it, several times, to make sure that the vibrato oscillator has sufficiently high gain to permit self-starting. Do not increase the value of VR2 more than absolutely necessary. Check that the depth of vibrato control works by varying VR3 up and down while monitoring the degree of frequency shifts at Tr2. If you do not own an oscilloscope there is no need to worry. An alternative method of testing will be given.

To complete assembly of this stage, insert DIL sockets for ICs 2 to 6 and double check that all the highlighted top side links are in position. Tick them off on the drawing as you check them! Before inserting these integrated circuits, connect up the power supply and loudspeaker but do not apply power yet. Taking great care to avoid static electricity (do not wear nylon or synthetic soled shoes. Touch your

## components

Power Skppy

| $\mathrm{C}_{4} 4$. | 4730\% 25 V |
| :---: | :---: |
|  | zgoustasV |
| c44. | 418007\% 4V |
| C4\% | 430nF |
| C48 | 470 nF |
| c4. | 4700F |
|  |  |
| C50 |  |

C11 470nF
$\mathrm{C} 42,10 \mathrm{nF}$
IC20 LM380
14-pin DIL socket. $6-25 \mathrm{~mm}$ Jack socket for external loudspeaker: 5 -pin $180^{\circ}$ DIN socket for phono output.

Tone Generators

finger on something that is earthed while touching one of the chassis tag pins of the board with the other hand) insert ICs 2 to 6 the right way round. Switch on and proceed to test for the presence of all the tones.

## Testing without instruments

Use the power amplifier as a test unit by injecting signals into it via the 'Phono Out' pin through C41. Cut a length of insulated wire and connect a crocodile clip to one end. This clip should be connected to the phono out pin using the lose end as a test probe.

First, hold this end on pin 8 of IC2 where the tone of top C should be present. With VR3 set to minimum try adjusting VR1 when the frequency of this note should change. Leave VR1 in the middle of its range and then increase VR3 to maximum. Probably, the vibrato oscillator is not oscillating so you will hear no change. Increase the value of VR2 until the onset of vibrato occurs. Set VR2 just high enough so that the vibrato always starts at switch on. Check that VR3 controls the depth of the vibrator.
Move the test probe to pin 3 of IC3 to monitor C, an octave lower. If not, check that the longer link to


Fig. 6: Tone generator and divider circuits. When testing this section, take special notice of instructions in the penultimate paragraph (over page).


## THE RADIO, TV AND AUDIO TECHNICAL REFERENCE BOOK

## S. W. Amos <br> Published by Newnes-Butterworths <br> 1172 pages $£ 24$

This is an unusual reference book, to say the least; to my knowledge, this is the first time that so many electronic and related disciplines have been gathered together within one volume. There are no fewer than 35 separate sections written by 31 authors covering everything from acceptance tests to zinc-silver oxide batteries.

The book weighs heavily towards technicians and all those who deal in the practicalities of a whole wealth of electronic hardware. To this end mathematics, although evident, take an intentional second place to photographs and drawings (there are over 1200 of them). Curiously, circuit diagrams do not feature as prominently as one would expect; where included, they are mostly notional although there is no real shortage of production equipment examples.

Taken generally, the contents indicates a slight degree of bias towards consumer electronics. For instance the sections on colour TV, taken together, are as specific and detailed as more single minded books dealing purely with that area. This book quotes the theory of the three systems (PAL, NTSC and SECAM) and enlarges on the practical aspects of the PAL system (as would be expected of a UK publication). It offers a very creditable description of tube types and convergence problems associated with the various gun arrangements. It details decoder operation and generally provides enough information for CTV line-up and test. As a slight contrast, the section on professional sound recording is a good deal thinner and generally fails to do this important topic justice; however most people who read this review won't be unduly perturbed.

The Technical Reference Book offers easy explanations of the simplest of circuits wnile wading in to far more difficult concepts in others. For instonce, it seems reasonable to suppose that most people who will use this book would be perfectly familiar with all the common amplifying configurations of a bipolar transistor. However, chapter 8 provides all the notional diagrams which will be instantly recognisable to all those who got past their first term at college. They don't need to be told about 'common base stage of amplification with stabilisation by potential divider and emitter resistor'. Similariy, the chapter on digital techniques was so facile that its author should have stuck to sucking eggs. . . . If a subject cannot be tackled at a reasonable level within the limited space allotment, then it should not be tackled at all. However, such criticisms are local and shouldn't disguise the value of other chapters and sections.

There are many chapters of interest to radio amateurs of which much material comes from personnel within the professional broadcasting organisations. For example, there are carefully explained descriptions and operating principles of all kinds of antennae including the slot variety; this topic has an adjunct of beam steering by phased
array which should hold much interest for G8's and the like.
Transmitters of all forms are dealt with at length; although broadcast installations are given more space (as they should be in a book of this type) there are specific sections on both amateur sound and television transmitters. Since amateur and professional areas require the same techniques and technicalities, all parts of the book are of interest, if not directly applicable.
It is quite impossible to do the Radio, TV and Audio Technical Reference Book justice in the space of one short review. It is much to the credit of edltor S. W. Amos to have organised so many contributors to cover so many topics under one edition. Even more noteworthy is the uniformity with which he has done so. There are wrinkles; several subjects overlap quite noticeably in places while others could do with greater coverage. However, the overall balance leaves little room for criticism. Although the price is high at £24, it seems worth the money, particularly if your technical library only extends to The Foundations of Wireless. The Technical Reference Book should only be purchased by those with a reasonable grasp of electronics and will be of greatest value to technicians and engineers.

Frank Ogden

JUBILEE Part 2 continued from previous page.
the left of IC3 is properly soldered in. Check at pin 4 of IC3 to hear the same note, yet another octave lower. Pin 5 of the same IC produces Bottom C. Carry out the same checks for the other notes of the scale referring to the pin numbers shown on the schematic of Fig. 6. You should end up with all 37 notes sounding clearly.

If all the notes are present at the outputs of IC2 but fail to appear in lower octaves, the fault almost certainly lies in the links; you may have bridged a couple of conductors with a solder blob. Do not forget that it is easy to leave out a soldered connection on the pins of the DIL sockets.
You are likely to experience some irregularity in tone generation at this stage. This is because the master oscillator operates from the +12 V stabilised supply (to ensure frequency stability) but the top octave generator and subsequent dividers operate from the unregulated +15 V rail. The latter power rail is not fully loaded at this stage of construction (ICs $7,8,9,13$ and 14 are not yet inserted) so its voltage will be higher than its design level. This temporarily reduces the input logic swing to the noise threshold. Some of the notes may 'burble' or sound irregular as a result.
If this situation occurs, you can compensate for the lack of loading from the omitted ICs by temporarily connecting a 270 ohm 1 watt resistor across the +15 V rail and ground (ie between the positive end of C45 and one of the pins which are designed to go to the chassis tag). This resistor will draw approximately the same current as the missing ICs and reduce the nominal +15 V rail to its design level.
Do not forget to remove this resistor when you come to insert the missing integrated circuits.

Before progressing with next month's instructions, it is wise to remove ICs 2 to 6 and put them back into their protective packing. To avoid damaging their pins, use a small screwdriver to lever them out of their sockets-a little bit at a time at each end.

Next month: Construction of the melody, rhythm and accompaniment sections.

## components


the positive line in the receiver and thus the drain voltage, without the need for other coupling circuits.
The 9 V positive line is taken from the receiver positive line, so the preselector needs no separate on/off switching.

## Layout

The panel is a flanged $4 \times 4$ in "universal chassis" member, and the variable capacitor VCl is mounted centrally, as in Fig. 9. Three short 4BA bolts are used here to avoid damage to the capacitor plates. The 2-pole 2 -way switch S1 is shown out of place, to clarify the wiring to it.
Other components are mounted on perforated board ( $0 \cdot 15 \mathrm{in}$ matrix) about $31_{2} \times 2 \mathrm{in}$. This is fixed to the bottom of VCl by running a 4BA bolt into the threaded hole. Washers are necessary between board and capacitor, and again ensure that the bolt does not project into the component and touch the internal plates. The few connections above and below the board will be seen from Fig. 9.

## Using the Preselector

The lead from 40673 drain to the aerial input circuit of the receiver should not be longer than necessary, and the aerial lead to the preselector is kept away from it. To simplify tuning, set the coil cores so that movement of VCl approximately matches that of the ganged tuning capacitor in the receiver. VCl is rotated to peak up wanted signals, but not those which constitute second channel interference. The latter signals will be at a point $3 \cdot 2 \mathrm{MHz}$ higher in frequency than the wanted transmissions, with the receiver having an IF of $1 \cdot 6 \mathrm{MHz}$. For a receiver with 470 kHz IF, offending second channel signals will be 940 kHz higher in frequency than wanted signals.

If neither the preselector nor the receiver have a metal case to provide screening, they need to be kept slightly apart, so that instability is not caused by coupling between the preselector and aerial coils. If trouble of this kind arises it is readily identified as whistles, and oscillation will be heard when the preselector is tuned to the same frequency as the receiver. Some screening between the units can be provided by sticking a piece of aluminium foil to the side of one of the cabinets, as mentioned in Part 1 and shown in the photograph on page 339 of Part 1. This foil must be connected to the chassis with a short lead.


The finished preselector. If it is used with a different receiver an internal 9V battery can be incorporated, with an on-off switch on VR1.

Remember that the capacitor Cl in series with the aerial, Fig. 1 in Part 1, must be removed from circuit when using the preselector otherwise the 9 V supply to the preselector will be interrupted. Even with a short aerial, strong signals can cause overloading, unless it is remembered to turn VR1 back as necessary.

The cabinet shown used 6 mm plywood sides, approximately $41_{2} \times 4^{1}{ }_{2}$ in in size, secured to the flanges present on the panel by means of nuts and bolts. The back is $4 \times 4$ in fitted with woodworking adhesive. Smail screws hold the top in place.

## Frank Henil G3GSW:

The sudden death of Frank Hennes G3GSW: on July 27 st has depriwed the a mateur Thio: \& vix Mateit of a distinguished
 1Hy an, theskiliphands. Frank's voice was heard by milliots of people every week When he presented the BBC. World Service prognamme, World Radio Club. Through Out his brodeasting life. Frank intervieved thousands of people for both radio and television and each item carried his owy special brand of humour. and that obvous personal care which was the hallmark of all lis efforts. It was great to work at he side of Trank, because, whatever the subject whetlie, he was serious or sport: The that characteristic smile. he simply oozel confideriee which was only domi nated by his sincerity
Que lune ssie (p, 07) carried a picture of Frank witervieving Lord Wallace of Costany, President of the RSGB, at the 500 th edition of World Radio Club, and, although his nultitude of friends will miss hin, Frank Henils will always be remem. bered as the one off' jovial individual portrayed in that picture.

Ron Ham BRS15744

## Redundant soldering iron

Broken track on a PCB? If only I could get that piece of glass to conduct. Two problems amongst many that can now be overcome with the careful use of a paintbrush. It sounds strange I know, but after five years successful use in industry a conductive paint is being made available to the general public. Called Elecolit 340 it is a pure, silver filled, electrically conductive acrylic paint which forms a tough film with

good adhesion to ceramics, glass, rubber, plastics, and most plastic films.
Typical applications include RF shielding, component lead termination, prototype PCB manufacture, PCB repair and a novel, but very useful idea, to repair the rear window demister of a car if a track gets broken. Elecolit 340 can be applied by dipping, brushing, silk screen or roller and will dry tack free in free air in 15 min . For the DIY reader the smallest quantity available is in 3 g bottles which sell for $£ 2 \cdot 70$, while for the person who thinks he can use 500 g he will have to part with £161-60!
Industrial Science Ltd., Leader House, 117-120 Snargate Street, Dover, Kent. Tel: 0304202656.

## Single board EM Tuner

Recently introduced by Ambit International, the 7122 AM tuner Module is suitable for three stage varicap tuning or for crystal controlled operation in the range 175 kHz to 30 MHz . Featuring the HA1197 IC as the heart of the package, input sensitivity is claimed to be $4-6 \mu \mathrm{~V} @ 1 \cdot 6 \mathrm{MHz}$, and $8-10 \mu \mathrm{~V}$ @

## Iron plug

Contrary to another heading on this page, there really is no substitute for a good soldering iron for the man who dabbles in electronics. A new one recently launched on the UK market is made by that "unlikeliest" of electronic firms-The Rawlplug Co. Ltd. Conforming to BS3456, this new 25W iron comprises a solid copper bit that is iron coated, and a double shaft consisting of a ceramic inner shaft and a stainless steel outer shaft for strength. The former has a breakdown voltage of $1,500 \mathrm{~V}$ AC.

Priced at $£ 3 \cdot 80+$ VAT it can be obtained from the usual outlets or from The Rawlplug Co. Ltd., Rawlplug House,

21 MHz for a weighted $\mathrm{S} / \mathrm{N}$ ratio of 26 dB .

Referring to the circuit, Cc value will vary with selected frequency range although $8 \cdot 2 \mathrm{pF}$ will suit most applications. Rif is nominally 270 ohms, although IF gain can be increased by reducing this value and the AGC reaction time constant capacitors, Cagc are $3 \cdot 3 \mu \mathrm{~F}$, although smaller values will speed up reaction time. The entire circuit is housed on a single PCB measuring approx. $90 \times 65 \mathrm{~mm}$ and will operate from a minimum supply of 12 V if the MVAM115 diodes are used. However in strong signal applications, and to minimise the effects of tuning voltage drift, the MVAM125 diodes may be used with a corresponding increase in voltage to 25 V . Priced at $£ 9 \cdot 00$ for the kit and £11.75 for a built and tested model, units can be obtained from Ambit International, 37 High Street, Brentwood, Essex, Tel: 0277227050.


London Road, Kingston upon Thames, Surrey. Tel: 01-546 2191.


## Boarded up

Trust the Yanks to come up with something just that little bit better, a little bit more sophisticated, and a little bit more expensive! Modules that accept components without the need for any soldering have been with us for some time now, and as a form for "breadboarding" they are invaluable. Now from America, but hopefully soon to be distributed in the UK comes the QT System which comprises sockets and bus strips which can be easily locked together in any pattern to form breadboards of any shape or size.

The sockets and bus strips will accept DIL IC's ( 6 to 40 pins), TO5s, diodes, resistors, capacitors, transistors, and just about any other discrete component with lead diameters between 0.015 and 0.032 inches. The contacts are solid silver-nickel alloy with a contact resistance of 0.4 milliohms and spaced 0.1 in apart. Each socket is made up of rows of 5 contacts, connected transversely at the back, while each bus strip has two rows of contacts connected lengthways. Connections between compo-
nents are made with standard solid 23 SWG wire.

Called the Protoboard 6 the basic kit contains one QT-47S socket ( $5 \times$ 94 contacts), two QT-47B bus strips ( $2 \times 40$ contacts), four 5 -way terminal posts, a metal earth and base plate with rubber feet and all the necessary assembly hardware. In all, one unit
measures $6 \times 4$ inches. Other sockets available range between $5 \times 118$ contacts and $2 \times 30$ contacts. Price for one PB-6 is $£ 10 \cdot 45$.
Further information including catalogue and UK price list available from Continental Specialities, 44 Kendall Street, PO Box 1942, New Haven, Connecticut 06509, USA.


## A case for sawing



In the construction of a project, there comes a time when you just have to put down that soldering iron, and get on with a bit of cabinet construction. It's a bit of a chore, but if the finished article is to look as well as it performs, then considerable care must be taken when measuring, drilling and cutting. To this end, Abrasive Tools Ltd., are marketing five packs of different size tension files (to the laymen-sawing wires) that will fit most all-metal 6 in junior and 10/12in Senior hacksaws. Prices range from 50p for a pack of five 6 in files to 70 p for a pack of three 12 in files. All are plus VAT.

Abrasive Tools Ltd., Abrium Works, Colne Road, Twickenham, Middx. Tel: 01-894 1273.

## 60 pages of data

The instrument Division of Gould Advance Ltd., has produced a new 60 page data book for 1977-78. The book gives full details of the entire range of oscilloscopes, digital voltmeters, timer-counters, pulse generators, signal generators and recorders and digital multimeters.
Gould Advance Ltd., Roebuck Road, Hainault, Essex.

## You can't resist this

The prolific home constructor or service engineer can now buy his resistors in quantity from Home Radio in the form of an Engineers Resistor Pack. Values from $4 \cdot 7 \Omega$ to $1 \cdot 5 \mathrm{M} \Omega$ are separately packed in individual clear plastic tubes and there are additional resistors of the more popular values making a grand total of 1600 in all. The resistors are all $5 \%$ tolerance and $\frac{1}{8}$ or $\frac{1}{4} \mathrm{~W}$ rating. Each tube is clearly marked and the backing cards retaining the tubes can be left in the original box

supplied to give quick access to any value of resistor required. The Pack costs $£ 17.50$ plus $12 \frac{1}{2} \%$ VAT plus 85p post, packing and insurance.

Home Radio, 234 London Road. Mitcham, Surrey CR4 6HD.

# DESIGN YOUROWN PROJEGTS 

This series describes how to design experimental projects for yourself. Each month we hope to set ourselves a reasonably simple problem, produce a set of specifications to which we want our circuit to conform, and design an appropriate circuit. We will give an account of the possibilities we explored and the thought processes we used to arrive at the final circuit, together with details of how we calculated (or guessed!) component values and why we chose particular component types. Last, but not least, we will provide an honest report of what happened when we built our circuit.

Toby's cassette tape recorder is normally run from the mains but he wanted to use it in his car without having to provide vast quantities of U2 batteries! The manufacturer had kindly provided a socket on the back marked "6V-DC." Hence this month's design example will be a power supply to convert the voltage available from the car battery to a 6 V supply for the cassette recorder.

## Specification

A plate stuck to the back of the recorder has " $3 W$ " printed on it, and whilst this seems quite a lot we had better allow for 0.75 A of current to be drawn as a maximum. Now the voltage of a car battery can vary quite a bit especially if the apparatus which controls the charging is somewhat dodgy! So, to be on the safe side, perhaps we had better cater for a car battery range of 11 to 18 V . As regards regulation of the output we shall specify that the voltage should remain between 5.5 and 6.5 V for output currents of 0 to 0.75 A . We shall have achieved our aim if we can produce a circuit which will perform to these criteria, so let us now turn to the design of the circuit.

## Design

We are going to require some component with which to stabilise the output voltage and a zener diode is the obvious choice. The simplest circuit using such a component is shown in Fig. 1. 6.2V is the nearest standard value of zener we can get to 6.0 V , although there is generally a $\pm 5 \%$ spread from the nominal value, anyway. Resistor R must take 0.75 A when the input voltage is 11 V so its maximum value is $\frac{11-6 \cdot 2}{0.75}=6 \cdot 4 \Omega$. At an input of 18 V it will be forced to take a current of $\frac{18-6 \cdot 2}{6 \cdot 4}=1 \cdot 84 \mathrm{~A}$. This means that $R$ will have to have a power rating of $V \times I=$ ( $18-6.2$ ) $\times 1.84=22$ watts and the zener will need a rating of $6.2 \times 1.84=12 \mathrm{~W}$ (when Iout $=0$ ). Well, 20W zeners are available but they are pretty expensive and the available values seem to start at above $6 \cdot 2 \mathrm{~V}$, but 22 W resistors are a bit excessive.


Fig. 1
No, this is getting out of hand, so we are going to have to use at least one active component, so let us look at the possible ways of doing this.

The easiest way is to use a shunt regulator. Note that the zener diode by itself is effectively a regulator as well, it works by shunting excess current away from the load to keep the output voltage down to the required level. The basic circuit of the single transistor shunt regulator is shown in Fig. 2.



Fig. 2: Use of a shunt regulator transistor to relieve the load on the zener diode.

This has taken a lot of pressure off the zener and shifted it to the transistor but the poor resistor is going to have to dissipate the same power as before. This is usually the problem with shunt regulators, inasmuch as excessive amounts of power are dissipated in the shunting element and associated components.
The other principal type of regulator is the series regulator whose circuit is shown in Fig. 3.
We have now used the simple shunt regulator ( Rl and the zener) to control the base current of Tr . The resistor no longer has to pass the output current, which is handled solely by the transistor. This now has to pass only 0.75 A as opposed to the maximum of $1 \cdot 84 \mathrm{~A}$ to which it was subjected before. This means that we have managed to reduce the dissipation from 22 W to 9 W which is more reasonable.


Fig. 3: The alternative series regulator circuit.
Nonetheless, 9 W is quite a lot and we are going to need a pretty hefty transistor here. One of the commonest NPN types is the 2N3055, but is this going to be OK in this circuit? The catalogue happily tells us that the maximum allowable power dissipation is 115 W at $25^{\circ} \mathrm{C}$ but do not be mislead by this. The figure quoted is for a case temperature of $25^{\circ} \mathrm{C}$ and if you start to dissipate 9 W in the transistor without some means of conducting all this heat away the case temperature is going to rise very quickly! For example, a 2 in square finned heatsink is quoted in the catalogue as having a thermal resistance of $6^{\circ} \mathrm{C} /$ Watt from the transistor to the temperature of the surrounding air. This means that if we are dissipating 9W the case will be at least $6 \times 9=54^{\circ} \mathrm{C}$ above ambient temperature. All in all you will almost be able to fry a small egg on the heatsink and as the temperature of the lump of silicon inside the transistor case will be even higher there is not all that much to spare.

We can actually let the junction temperature creep up to $200^{\circ} \mathrm{C}$, so we should be OK as long as we heatsink the transistor adequately. The simplest solution seems to be to make the unit in a diecast aluminium box and to bolt the transistor to the outside, plus silicone grease and an insulating kit if necessary. The box will help to make the unit nice and robust which is a decided advantage in this case.

Resistor R2 is included just to draw a small current when the load is not connected. By doing this we do not have to worry about the various esoteric effects which are prone to occur at very low currents and high junction temperatures, and which might conceivably cause loss of regulation of the output voltage. We could easily do without R2 but it is good practice to put it in. A value of $2 \cdot 2 \mathrm{k} \Omega$ which gives a minimum load current of 2 mA should be quite adequate. The zener value will now have to be $6 \cdot 8 \mathrm{~V}$ to allow for the 0.6 V base-emitter voltage drop in Trl. The circuit so far is shown in Fig. 4.


Fig. 4: The practical circuit so far with component values.
What about the value of R1? This depends on the value of base current needed to drive Trl. The minimum gain of a 2 N 3055 for a collector current of 4A is given as 20. This figure should do for our purposes as the gain should be substantially the same at 0.75 A . Hence we are going to have a maximum base current of $\frac{0 \cdot 75}{20}=38 \mathrm{~mA}$. What about leakage current which may be quite large in a hot power transistor? Well, this is going to flow in the opposite direction to the rest of the base current and so tends to reduce it so we are OK here. To supply 38 mA at the minimum input voltage we need R1 to be at most $\frac{11-6 \cdot 8}{38}=110 \Omega$, so $100 \Omega$ should supply enough extra current for the zener as well.

When the input voltage is 18 V R1 will be passing 18-6.8
$100=112 \mathrm{~mA}$ and, if we take the worst case of all, this current going through the zener so then the maximum dissipation in this component will be $6.8 \times 0.112=0.76 \mathrm{~W}$. The dissipation in R1 will be $\mathrm{I}^{2} \mathrm{R}=(0.112)^{2} \times 100=1 \cdot 25 \mathrm{~W}$, hence a 1.5 W resistor for R1 and a 1 W zener should be quite adequate.

## components



R2 $2.2 \mathrm{k} \Omega \frac{\mathrm{L}}{4} \mathrm{~W}$ R3 0.688 .1 W D1 6.8 V 1 W zener diode Tr2 BFY51

## Miscellaneous

Diecast box $4 \frac{1}{2} \times 2 \frac{1}{2} \times 1 \mathrm{in}$. or similar. Terminal block. Stripboard about $3 \frac{1}{4} \times 1 \frac{3}{4} \mathrm{~h}$, Insulating kit for 2 N 3055 . Heatsink for BFY51.

At this juncture it was suggested that we might add another transistor to end up with either of the circuits in Fig. 5, circuit (b) being the more sophisticated. We would like to avoid having to use high wattage resistors and zeners but to do this the services of


Fig. 5: above, a second transistor can be used in either of these circults. Circuit (a) was chosen.
Fig. 6: below, shows practical values for Fig. 5 plus a smoothing capacitor across the output.



Fig. 7: The final circuit which Includes an overload sensing transistor Tr2.
The principle of simple protection is that you sense the current being supplied with a low value resistor and turn on a transistor when the voltage drop across the resistor reaches about 0.6 V . Of course, you do not drop the output voltage by fighting with the emitter of the power transistor. A rather easier way is to drag the base of $\operatorname{Tr} 1$ down to earth by shorting out the zener and passing a rather higher current through R1. This can be achieved by an NPN transistor with the current sensing resistor connected between its base and emitter and its collector connected to the base of the 2N3055.

Where does the current sensing resistor go? One of us, exhibiting a rare and temporary aberration, drew in the resistor at the point marked X in Fig. 6, but Bob does not want us to say who it was. The point is that the zener is controlling the voltage between the emitter of $\operatorname{Tr} 1$ and the earth line and sticking a resistor between the zener and the negative output will increase the output voltage by up to half a volt at the maximum current. No, the resistor will have to go right back at the input, the fact that the

 used by the badeberoon a piece of stripdotat:
another transistor are required. In the end the virtues of simplicity won the day and we pushed onwards with our original design which was now at the stage shown in Fig. 6. We have added the capacitor across the output to get rid of any ripple or anything else lurking on the car electrics.

## Protection

Now, what can go wrong? Well, if one of us contrives to short-circuit the output terminals (by a well known law of nature this is virtually certain to happen sooner or later) then the power transistor is not going to survive for more than a second or so. Since our supply of 2 N3055s is strictly limited we would like to incorporate some form of short-circuit protection.
current being sensed now includes that being supplied to the zener does not matter much, since this extra current is a lot smaller than the current being supplied when limiting is taking place. We now have the circuit in Fig. 7. A suggested layout is shown in Fig. 8.
We require the base-emitter voltage of $\operatorname{Tr} 2$ to reach 0.6 V for currents of about 0.85 A , thus giving R3 a value of $0 \cdot 7 \Omega$ but $0 \cdot 68 \Omega$ IW will have to suffice. The power dissipated in $\operatorname{Tr} 2$ will be highest when it is fully on and Vce is around 6.8 V . Ic will then be 112 mA (for an input supply of 18 V , as before) and the power dissipated, given by Vce x Ic, comes to $0 \cdot 76 \mathrm{~W}$. The maximum dissipation of a BFY51 is given as 0.8 W so if we use one of these with a clip-on finned heatsink or something similar we should just continued on page 454


A REVIEW OF RECENT DEVELOPMENTS
In general, the author does not have any more information on products than appears in the article.

## The CB Saga!

Citizens Band arguments still smoulder, but while potential users and opposers debate the finer points, the manufacturers are steaming full speed ahead. One possible advantage in the UK is that when CB does come, it will bring with it some very sophisticated pieces of equipment.

One manufacturer has launched a CB system, transmitter and receiver which uses microcomputers in both! The microphone looks rather like a small pocket calculator. The mike is housed in the uppermost portion of the hand held unit. Directly beneath this is a 5-LED readout and underneath these are all the small push buttons. Channels are selected by simply pressing the buttons and the readouts confirm the channel.

It is interesting that the internal microcomputer is used to electronically control a CCD (charge-coupled device) filter. It does this in such a way that there is never any need to touch anything to "resolve" SSB signals. Bandwidth, bandpass and automatic frequency locking are all achieved by the microcomputer.

No cheap gimmicks either. The receivers are triple-conversion types with a few very mouth-watering advantages. Things like fully automatic scanning of all channels, automatic and continuous measurement of the VSWR and, if the VSWR gets too high, the set will automatically shut down. Automatic level and gain are other convenient features.

Power for these transceivers is 12 W pep with the option of $4 W$ of AM. In the SSB mode, upper and lower sidebands are selectable.

In the USA, the launch price of the mobile station is around the £200 mark.

## 10 min. a day

I was interested to hear about a Japanese company in the watch making game. They launched an electronic watch with a conventional face (i.e. hands plus day and date) but with great cunning this company also designed the face of the watch to take in eight silicon solar cells. It is reckoned that a battery will last at
least three years and that only ten minutes of light on the cells every 24 hours is sufficient to keep the battery charged. One of the solar cells is connected in reverse to prevent reverse current from flowing when the other "working" cells are not receiving any light. The approach seems to be a success; over 50,000 of these watches have been sold in only eight months.

## Thang a TV on your wall

Flat screen televisions took a step nearer commercial use recently when a Japanese company released details of some work being carried out with liquid crystal TV screens. This one is a six-inch screen with 8,938 elements on it. The device does work and has been demonstrated in a television receiver application.
The liquid crystal screen is driven by CMOS, and drivers plus screen take only 10 mW . The remainder of the set consumes only 5 W and this includes the audio stages. While some work still remains to perfect this approach, the company is already talking about a pocket television receiver which will be no larger than a small pocket calculator. This is envisaged within the next three years.

## Sad ism'佔?

Be kind to tin! Did you know that metals "cry" when stressed? Well they do, due to molecular dislocations, micro-cracking and molecular deformations. With special equipment you can hear these noises. German workers are now using sound emission analysis to obtain advance warning in things like nuclear reactor vessels. The characteristics of a particular "cry" tells the workers just what's happening inside the structure of the metal; whether its just a small stress, or if there is some more serious damage or weakness, such as a crack starting.

I must confess that banging the Ginsberg begging bowl brought no sobs of distress from the metal-only from my bank manager!

## This takes the biscuit!

Just what you've always wantedan electronic biscuit checker! Not for the housewife, perhaps, but the biscuit manufacturers have to be a bit careful. After all, they can't send out one fat biscuit and three skinny ones, can they? It's all done with mirrors, plus some electronics, of course. The system projects a slit of light onto the surfaces of the passing biscuit. The scattered light is scanned by a photomatrix and the resultant information is used to adjust the temperature of the oven. The system can measure to within 0.1 mm accuracy, but the designers claim than an improved version will allow accuracies down to 0.01 mm .

## Soon-mass Xeray in space

I shudder to imagine the enormous complexity of the latest space shot planned by the USA. Apparently they are to launch a half-ton X-ray detector. The earth's atmosphere prevents this type of detection efficiently from the ground.
Called Laxray, it will be able to pinpoint X-ray sources to within $0.1^{\circ}$ and will do this for sources as far afield as a billion light years. The scientists expect Laxray to find many thousands of X-Ray sources with names like quasars and black holes. Makes that S-DeC project you just built seem a bit tame, doesn't it?

## Even cheaper by Xmas

A late news flash for games chip addicts. A very big chip manufacturer in the USA is to launch a chip which should put games systems in the "less than ten dollars" market by this Christmas. Further, the manufacturer reckons to add to this chip family and next year the aim is to bring the total system cost down to a miserly five dollars. As the electronic canary said, "Cheap cheap".
Cimberz

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# RUDIO LEUEL IDDICATOR <br> PEAK PROGRAMME RESPONSE WITH LED DISPLAY 

This peak reading audio level indicator may be incorporated into a tape recorder or the modulating system of a transmitter, or used to give visual indication of the output of an audio amplifier. The indication is by means of four light-emitting diodes which light in sequence as the input increases. These save the high cost of a moving-coil meter, while still giving an easily assimilated display. A more complex circuit is necessary but the components are inexpensive.

The design responds to the peak-to-peak value of the signal. Thus, both positive and negative peaks are taken into account, which is a slight advantage with asymmetric waveforms such as male voices.

## Circuit description

The circuit is shown in Fig. 1 and 2 where VR1 is used to set the sensitivity. $\operatorname{Tr} 1$ and $\operatorname{Tr} 2$ form an amplifier with a voltage gain to audio signals of about 50 times (R4/R3). The function of C2 is to prevent the amplifier from oscillating at a supersonic frequency because of stray capacitive feedback. The amplified signal is applied via R6 to a peak-to-peak detector comprising C5, C6, D1 and D2. R6 gives a charging time constant of a few milliseconds, the accepted charging time constant for a peak programme meter being 2.5 ms . This period may be shortened by reducing the value of R6, or omitting it completely. The discharge time constant is about one second ( $0 \cdot 22 \mu \mathrm{~F} \times 5 \cdot 6 \mathrm{M} \Omega$ ).

Tr3 and Tr4 are connected as a Darlington emitterfollower so as to prevent a very high resistance to the rectifier and preserve the full time-constant. R7 and VR2 apply a forward bias to overcome the cut-in voltages of diodes D1 and D2 and of the base-emitter junctions of $\operatorname{Tr} 3$ and $\operatorname{Tr} 4 . \mathrm{C} 4$ is an audio bypass.

The full peak-to-peak voltage thus appears at $\operatorname{Tr} 4$ emitter. In the absence of any signal, this voltage is zero and $\operatorname{Tr} 5, \operatorname{Tr} 7, \operatorname{Tr} 9$ and $\operatorname{Tr} 11$ are all non conducting. $\operatorname{Tr} 6, \operatorname{Tr} 8, \operatorname{Tr} 10$ and $\operatorname{Tr} 12$ are all in saturation by. passing the current supplied by R20 from any of the light-emitting diodes. If the input voltage is gradually increased, Tr5 will begin to conduct when the rectifier output reaches approximately 0.5 V . Shortly afterwards, it will conduct sufficiently to shunt the current provided by R10 away from the base-emitter junction of Tr6. Tr6 will cease to conduct and LED1 will light.

When the rectifier output exceeds approximately ((R12 + R11)/R11 $\times 0.5 \mathrm{~V}$, or approximately $5 \mathrm{~V}, \operatorname{Tr} 7$ will conduct and LED2 will come on, Similarly, LED3 will light at 10 V and LED4 will light at 15 V . With VR1 set for maximum sensitivity, the signal voltages at which the LEDs conduct are roughly $15 \mathrm{mV}, 100 \mathrm{mV}$, 200 mV and 300 mV p-p. The sensitivity increases with increasing temperature, but only by about $3 \%$ for a rise of $10^{\circ} \mathrm{C}$.


## Components

The prototype used selected transistors from untested ones which were bought for 50 p per 50 ( $80 \%$ good). They are similar to ZTX108(NPN) and ZTX500(PNP). The eight transistors in the switching section may have an HFE as low as 50 and also low breakdown voltages. However, $\operatorname{Tr} 1$ and $\operatorname{Tr} 3$ should have high gains and low leakages.

When used with a tape recorder, it is a good idea to use green LEDs for the first three and a red one for LED4, as an overload indicator. VR1 is then adjusted so that the third LED lights at the optimum peak recording level. LED1 will be found to light almost continuously with normal programme input.

## Construction

Fig. 3 and 4 show the printed circuit layout. Singlesided Vero-pins were used for the power and signal inputs and for the LEDs. Soldering the LEDs directly to the pins is a convenient method of mounting them in a straight line. Alternatively, they may be situated away from the board and connected by cable.

## Setting-up

VR1 should be adjusted temporarily for zero sensitivity to avoid picking up hum and other noises. A sensitive test meter ( $50 \mu \mathrm{~A}$ FSD) should be connected across the base-emitter junction of Tr5. VR2 should be advanced from zero until the meter begins to register and then reduced slightly, Alternatively, VR2 may be advanced until LEDI just lights, and is then turned back a little.

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Fig. 1 (top): The circuit diagram of the Buffer amplifer and rectifer. The rectified peak to peak voltage appears at the emitter of Tr4.

Fige 2 (middele the voltmeter section. 7 Tr, 7,9 and 17 tarn on successively with a rising voltage an the common fine from Tr4 enitter, for satisfactory operation the supply vollage must not be less thân 422 V or more than 330 V.

Fig. 3 (below): The PCB shown copper side actuaI size A ready made board may be obtained from the PW Readers PCB service ff required.

Fig. 4 (bottom): The component overtay for the PCB.

## components



VR1 should now be adjusted so that LED4, or LED3 if LED4 is to indicate overloads, lights at the desired maximum level. In a tape recorder, the unit must, of course, be connected after the recordinglevel control. For a stereo system, the best method is to have two separate indicators, but a single one can be switched manually between channels. The signal should be taken from points of low impedance relative to the input impedance of the indicator. This lies between $20 \mathrm{k} \Omega$ and $100 \mathrm{k} \Omega$, depending on the setting of VR1.


The current consumption of the circuit is approximately 25 mA from a 25 V supply. The display is bright enough to be clear in fairly strong daylight.



by Eric Dowdeswell G4AR

There seems to be little doubt now that we are starting to climb out of the doldrums of the last sunspot cycle. Activity on the 10 m band is increasing rapidly and it behoves all of us to get those 10 m beams dusted off ready for the fray. Many readers of this column will not have experienced 10 m in full cry and they are in for a very exciting time. Those with transmitting licences will find themselves working all over the world with the simplest of equipment and low power. However, because of the small physical size of 10 m beams a multi-element job is not a massive engineering undertaking. Not to get the gain necessarily but to provide some discrimination against the QRM!
The performance of a receiver at these frequencies should not be taken for granted. The older sets may be fine on 80 m but abysmal on 10 m . It is not a bad idea to make up a crystal-controlled converter for this band and tune the station receiver over a convenient part of the dial such as 4 to 6 MHz thus taking advantage of the facilities of the set. It is a fact that much of the reported lack of activity on the higher frequency bands is due not to conditions but to poor receivers!
Paul Bradbeer runs from one problem to another! In Braunton, Devon, he has got his exams over but now finds his summer job interfering with his listening activities! However, he seems to be putting enough aside to buy a decent receiver and a trap vertical aerial in the not too distant future. Can't be too bad! Although not a VHF man Paul was sorely tempted when he joined in the VHF NFD fun with his local radio club. In Holyhead John Higginbotham BRS36901 is determined to use his holiday to up his DX figures. In anticipation of the RAE result John has acquired an FR100B and FL200B. I do hope that you have been successful OM or you will have to put that TX into mothballs for a while!
"As you may know, we schoolchildren have five weeks off now for summer holidays, so I'll have a chance to do some 'night owling'," so says Michael Walker BRS38836 up in Leeds. He frequents the White Rose club there and expects to start an RAE course later in the year. Paul Pasquet and Iain Christie
jointly looked over the 15 and 20 m bands in Farnham, Surrey, and found fairly scarce ones like HH2, VP1 and KG4 but don't seem to have ventured to the 10 m band. In London W4 Denis Sullivan admits to finding the amateur bands a bit more interesting than the aircraft bands on HF. As ex-G2FCJ he couldn't resist the temptation to get back and has acquired a Trio QR666 used with "a bit of wire in the loft". He gave up his ticket in the '60s which seems rather a pity as he too comments on finding the 10 m band producing signals at 2230.

An interesting letter from Louis Meulstee PA0PCR who was in EI land on holiday as EI2VLB/M with his Heathkit SB104 and whip aerial on 10, 15 and 20 m . In Hastings 10 m was scanned by Brian Harrison producing C31, FM7, HI8 while 20 m turned up JW7 (Bear Is.), KG6 and VK9 (Christmas Is.). The neighbours of Dave Peck BRS37621 in Cambridge can now rest easy as he has cured the QRM he was causing them, by sprinkling around his RTTY gear an assortment of capacitors and chokes. Not much to report says Dave but he did copy a DM on RTTY via Oscar 7 on 2 m . Writing from Walsall, M. F. Wilson says he has been very active in DX-TV using the sporadic-E mode, "but now wishes to move into the world of amateur TV". If there are any enthusiasts in this field in the Walsall area he would like a note to 406 Sutton Road.

Brian Alderwick's letter was his first to this column but not the last, I hope. He has replaced his 9R59DS with a Yaesu FRG7, the former being relegated to MW DXing with a PW look aerial! At which it performs very well indeed, as I can personally testify! His brief $\log$ covers 20 m SSB only so I hope that you will not be averse to using the bandswitch now and again OM! Like 10 m ? Don't worry Brian you won't revert to the BC bands now that you have tasted our bands!
The $\log$ from D. W. Waddell in Herne Bay covered 10 to 80 m with a mixture of CW and SSB which is as it should be. That is the only way in which to get a good picture of what conditions are up to, enabling one to select the appropriate band that is active, without wasting time. Even if you do not have a couple of hours to sit down with the receiver a few minutes switching from band to band will put you in the picture.

CLUB NEWS. The Newport ARS meets every Monday at 1900 at the Community Centre, Brynglas, during the school term. New aerials being installed now include a TH2 for the HF bands with a 4 -over-4 slot for 2 m plus a 19 ft colinear for the same band. Forthcoming programmes include: September 12th, "Linear Amplifiers"; October 10th, "Oscilloscopes", both by GW3NWS. Interested? Then contact M. L.

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The Wessex AR Group will be pleased to see you at 1930 in the Club room at the Dolphin Hotel, Holdenhurst Road, Bournemouth, for the following: September 16th, "Working Maritime Mobile" by Bill Sykes G2HCG and Dick Weston G3RGJ. The AGM is on October 7th, but for more info contact Geoff Cole G4EMN, 6 St. Anthonys Road, Bournemouth. Tel: 20027.

## Log extracts

D. Waddell:- 80 m 9G1JX 20 m HH2MC KH6HC KL/AM SU1M1 VP8PL 5T5CJ 15m AP2P ST2SA TL8LE ZD8EW 8R1J 10m TU2FM ZP5SD
B. Alderwick:- 20 m CO2WK CX5BT HSIALG JY25DI VP2LCT 6Y5LB
B. Harrison:- 20 m JW7BK (Bear Is.) KG6BZ TJ1BB VK9XI (Christmas Is.) 3B8BZ 8R1Q 15m VP2MBB 9Y4R 10m C31NX FM7WE HI8EJH 9Y4AL
P. Pasquet and I. Christie:-20m C3INX KJ6BD TR8BA ZF1SV 5W1BF 15m CJ1EJ (Prince Edward Is.) FG7AS HH2MC KG4AN VP1WCS VP2MVP
A. Butler: -20 m CO2HZ JY25MB KP6BD
J. Higginbotham:- 20 m CR9AJ HK0QA (San Andres) KG400 KJ6BZ KX6DC TJ1BB TT8SM VS6DO 8R1W 15m CE3FI CX7BF PJ2FR VP2MVP
P. Bradbeer:- 15m JY9CR VP2MBB (QSL cards to VE3ECP) 9Y4R


## MEDIUM WAVE DX

## by Charles Molloy G8BUS

South East Europe is an area neglected by the medium wave DXer. Signals are not conspicuous, channels are often shared with western Europe, and the best time for DXing is in the early morning before broadcasting from the west starts up for the day, a time when the majority of DXers are in bed!

Greece is one of these neglected countries. It's time zone is two hours ahead of GMT and its main channel is on 728 kHz which is shared with East Germany, Austria, Spain and Portugal. It is possible though, to pick up a few signals in the late evening. Corfu is at reasonable strength on 1007 kHz after Hilversum signs-off at 2300 ( 2200 GMT in the summer) while Zakinthos is on 926 kHz at the same hour with a weaker signal, after the Belgian station on that channel has closed down. Athens on 1385 kHz has a night programme and a loop will reduce the QRM on this channel.

YENED is the Hellenic Armed Forces Network which operates a chain of low-power outlets that signoff at 2100 . Those logged by the writer are Kavala with 1 kW on 1355 and Serrae with 1 kW on 1301. The Pyrgos Broadcasting Station also run by YENED, has a night programme with announcements in English which is on 1349 kHz with 4 kW from 2300 until 0300 . QRM is usually troublesome on this frequency. Two out-of-band transmissions can be heard during the winter. Radio Ierapetra on the island of

Crete is on 1614 with a power of 250 watts and according to a letter from the station it is located at Ierapetra on the south of Crete, and signs-off at 2200 . The other is Amphissa on 1622 which closes down at 2100. Kalamata is listed as being on 1620 but has not been heard by the writer.
Mark Brighton writes from Benfleet in Essex. "A couple of months ago I began DXing on the medium waves, mainly because I have no radio with short waves". Shame on you Mark, medium wave DXers prefer their band to the short waves! Mark says his receiver in an ancient 5 -valve domestic superhet connected to 100 ft of wire strung between his bedroom window and the apple tree up the garden. He tried for CJON at 0200 but thinks he might have heard AFN Berlin on 935 kHz . AFN is on a 24 -hour schedule and could easily be mistaken for a North American. The majority of North Americans identify frequently by their callsigns and all are compelled by law to announce their identity on the hour and the half hour. Good luck with your DXing Mark, and stick to the medium waves!

Leslie Dewhurst writes from Leamington Spa to say that he has been a regular but silent reader of this column for the last twelve months. He had some small successes with North America last winter and he now has a Realistic DX160 receiver. He is preparing for the darker nights and is putting up a 100 ft long wire aerial with a screened downlead in order to get rid of local electrical interference. He purchased a notch filter kit and has collected the "bits" and pieces" to make a loop.

Leslie raises an interesting point when he asks if readers would give the times when reporting DX. He says "it would be a great help to beginners like myself, to avoid the frustration of searching when there is in fact no chance of success'. Well, readers, please help others by giving details of your gear, aerials and the date and time (GMT) of your loggings.

Loops are in the news again with a letter from Exeter where Richard Harding logged WINS on 1010 and WCBS on 880 both in New York and CHER in Sydney, Nova Scotia. Reception was in July between 0130 and 0430! The receiver is an Astrad VEf 17 with a loop measuring 60 in high and 47 in wide coupled to the receiver via a differential amplifier and an Aerial Tuning Unit. Richard asks; Do you know of anyone else who uses a rectangular loop and is there any limit to the size a loop may be.

A loop may be any shape that is symmetrical; square, rectangular, circular, diamond shaped or even triangular will do. Some DXers in the United States use a delta loop which is shaped like an isosceles triangle with the apex at the top and the opposite short side being the base. Yes, there is a maximum size and this would be a single turn loop which, if square, would have each side about 25 ft in length! This estimate is based on the rule-of-thumb measurement that holds in practice with smaller sizes; that it takes about 100 ft of wire to wind a medium wave loop.

The pick-up of a loop is proportional to the area enclosed by the windings and to the number of turns. If those two are multiplied together a figure will be obtained that will be a measure of the pick-up of the loop. For example, the figure for a 40in loop with 7 turns is 77 , for a 6 ft loop with 4 turns it is 144 and for an 8 ft loop with 3 turns it is 192. Clearly the larger loops pick up more signal but unless they are well made and rigid then the null may not be too good. The standard loop which is an optimum size
between pick up, convenience in use and sharpness of null is the square loop of 40 in side with 7 turns.

Does anyone use a Trio QR666 receiver for medium wave DXing? Richard Harding would like to know how this receiver handles on the medium waves and what stations have been logged with it. Can anyone help?

Following up a request last month from Derek Taylor for information about low power stations in the United States. "Broadcasting Year Book" obtainable from Broadcasting, Telecasting Building, De Salles St, Washington DC 20036, USA, gives the addresses, transmission times and names of the owners of all medium wave stations in the United States. The total number of stations, including a large number of 250 watt daytimers only, is in excess of 4000 and the cost of the Year Book is 20 dollars. In a subsequent letter Derek mentions the National Radio Club Domestic Log, which covers USA and Canadian stations. Stations are listed in frequency order and also by callsign. Details of addresses, transmission times, power used and other data are included in the $\log$ which is obtainable from N.R.C. Publications Centre, PO Box 401, Gales Ferry, CT 06335, USA at a price of 8 dollars to non members.


## SHORT WAVE BROADCASTS by Charles Molloy g8bus

The short waves are used for domestic broadcasting in tropical countries and in other parts of the world where the high level of static caused by thunderstorm activity makes broadcasting on the medium waves impracticable. The Tropical Bands, mainly 60 metres and 90 metres, are used during the hours of darkness, a time when propagation is by means of the $F$ layer, to provide a service for several hundred miles around the transmitter.

During the day, when the sky wave is absorbed in the $D$ and $E$ layers, a change to higher frequencies is required to maintain the service. Consequently, many broadcasters operate on a split schedule. They are on the tropical bands from sign-on until sunrise, then they move to the $49 \mathrm{~m}, 41 \mathrm{~m}$ or 31 m bands during the daytime and return to the tropical bands from sunset until sign-off. The night time signals travel considerable distances and if the DXer wishes to pick them up he should listen at a time when there is a path of darkness between the transmitter and his location (QTH).

## 60 metres $(4750 \mathrm{kHz}$ to 5060 kHz )

This is the main tropical band and worldwide reception is possible. At this time of year listen for:

4756 kHz Brazzaville, Congo. French. Close down 2300
4770
4780 Radio Bissau, Guinea. Portuguese. C.d. midnight

4815 Radio Ougadougo, Upper Volta. French. C.d. midnight
Cotonou, Benin. French. C.d. 2300
Port Moresby, New Guinea. Sign-on 2000
Radio N'Djamena, Chad. French. C.d. 2200 ABC, Brisbane. Sign-on 1900 (1930 Sundays)
Radio Abidjan, Ivory Coast. French. C.d. midnight Radio Yaounde, Cameroon. French. C.d. 2300

90 metres $(3200 \mathrm{kHz}$ to 3400 kHz )
3222 kHz Radio Lawa Kara, Togo. French. C.d. 2230
3227 Radio ELWA. Local languages. C.d. 2230
3232
Brazzaville. C.d. 2300
3330 Radio Rwanda. French. C.d. 2100
3380 Blantyre, Malawi. English. C.d. 2210

Programmes for DXers are a feature of many of the international broadcasts on the short waves. News items of interest to DXers, latest schedules, information about new stations, frequency changes, answers to technical questions from listeners, are the meat of these programmes and one station, Radio Nederland in Holland, even runs courses on various aspects of DXing.
'Sweden Calling DXers' has been on the air for many years. The English version, which is on Tuesdays, can be heard in Europe on 6065 kHz in the 49 m band at 1615, 1845 and 2115. Reception in parts of the UK can be difficult after dark but the programme is repeated on the medium waves on 1178 kHz at 2300 . The items broadcast are supplied by listeners who receive in return a copy of the script of the programme in the form of a weekly news sheet.
'DX Juke Box' is the title of the programme from Radio Nederland. It is on the air on Thursdays on two frequencies in the 49 m band. On 6045 kHz at 1415 and on 6030 kHz at 1845. A 'Communications Systems' course is currently included in this programme. The Swiss 'Short Wave Merry-go-round' appears on the second and fourth Saturdays of the month on 3985 kHz in the 75 m band, on 6165 kHz in the 49 m band and 9535 kHz in the 31 m band. Listen at $0705,1105,1320$, 1535 or 2105.

Although DX is generally accepted as an abbreviation for Distance, the suggestion put forward by Eric Dowdeswell in the August issue that 'Distance Xtra' would be better, touches the heart of the matter. There is a lot more to DXing than listening to distant stations. At the risk of trying to improve on this novel suggestion from the Assistant Editor the writer would suggest that by far the best definition of DX is the one put forward many years ago by the late G6QB who suggested simply that DX stands for Difficulty!

The DXer who owns an efficient communications receiver, lives in a country district free from electrical noise, and has an aerial farm, will have a different idea of difficulty than the owner of less pretentious gear who is surrounded by high-rise flats and has to use an indoor aerial! The feeling of achievement however will be the same when some difficult DX has been winkled out and that, really, is what the hobby is all about.

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An attenuator is a device, usually made of resistors, which will absorb electrical energy. If one is placed between the aerial and the aerial socket on a receiver it will cause a reduction in signal strength. This would seem a strange thing to do as usually the object is to collect as much signal as possible and then amplify it within the receiver so that weak signals can be heard. There is an exception though Overloading inside the receiver can occur when a long outdoor aerial is connected to a receiver that is designed to operate from a short aerial, such as a whip or a car aerial. Signals will spread out on the dial, spurious responses will occur along with whistles which will make it very difficult to listen to any but the strongest stations. When this occurs it will be an advantage to switch in an attenuator which will reduce the strength of the stronger signals, reduce overloading and make the weaker signals audible.
A simple yet effective attenuator is a $500 \Omega$ or $1 \mathrm{k} \Omega$ carbon potentiometer connected between the aerial and earth terminals of the receiver with the aerial going to the rotor. (See the article in this issue on a SW bands pre-selector.)
A variable capacitor in series with the aerial will also act as an attenuator. The capacitor should be mounted in a small box with an input socket for the aerial plug and an output lead with a plug on the end which is inserted in the receiver aerial socket. The variable capacitor is normally left with the vanes fully meshed, i.e. at maximum value, when it will have little effect on receiver sensitivity. When overloading occurs, the capacitor is adjusted until weaker but clearer signals are obtained.
Paradoxically, a capacitor in series with the aerial will, under some circumstances, peak up a signal. It has the effect of electrically shortening the aerial and can be used as a simple matching device. It is worth trying one in series with a long outdoor aerial especially when listening to the higher frequencies.
Radio Australia broadcasts to Europe in English daily between 0700 and 0900 . It can be heard on the 25 m band on 11740 kHz and on the 31 m band where it has moved back from 9510 kHz to 9570 . It has also been reported on 21570 kHz on 13 m though it is not clear whether this frequency is beamed on Europe. 'DXers Calling' is on the air on Sundays at 0840.

According to Deutsche Welle they are now using $6000,15320,15405$ and 17875 kHz from their relay in Malta, and 5960 , 5995, 6145, 9545, 9590, 9625, 9690, 11865 and 11970 kHz from the relay in Montserrat.

by Ron Ham BRS15744
Although sporadic-E is an annual mid-summer phenomenon its sudden and varying effect on radio signals still fascinates the seasoned enthusiast and excites the newcomer to VHF, and Saturday June 25 was a good example. For most of the day both the 4 and 6 m bands were in chaos. My frequent checks revealed strong vision pulses on Ch.R1 ( 49.75 MHz ) and very strong signals from some 40 east-European broadcast stations between 65 and 73 MHz , with only dipole aerials feeding my 770R and R216 receivers.

Reports on the various bands are welcome and should be sent direct, by the 15 th of the month, to:-
AMATEUR BANDS Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey KT21 2TW. Logs by bands, each in alphabetical order.
MEDIUM and SW BANDS Charles Molloy G8BUS, 132 Segars Lane, Southport, PR8 3JG. Reports for both bands must be kept separate.
VHF BANDS Ron Ham BRS15744, Faraday, Greyfriars, Storrington, Sussex RH20 4HE.

Frank Luman, Glasgow, reports that Band II was disturbed throughout the afternoon, "They came thick and fast" says Frank and while he was counting the Italian and Spanish broadcast stations he noticed that the BBC transmissions from Holme Moss on 91.5 MHz were often obliterated by the Italian Programme 2 station at M.Sambuco.

Alan Baker G8LGQ Newhaven, reports that around 1600 HG5AIR, on 2 m SSB, was heard in Haywards Heath and about the same time IT9 was received in GM. OE and SP were worked from G and G8HVD along with other stations in G, GW and GU successfully contacted 9 H 1 . Further sporadic-E disturbances affecting the 4 and 6 m bands occurred during the early mornings of July $1,3,8,9,11$ and 16 , at midday on the 8th, and the late afternoons of the 5th and 12th. The most intense of these took place at midday on the 8th when I counted 63 strong signals from east-European stations between 65 and 73 MHz , in addition to a variety of continental radio-telephone signals between 40 and 46 MHz . On the other disturbed days an average of 14 broadcast stations were heard in the 4 m band.

Frank Luman has recently purchased a Hallicrafters S36A receiver to extend his DX listening into Band I and has already heard TV sound carriers from both French and Spanish stations. John Branegan, Saline, Fife, has recently retired from the Navy and is in the process of building up his station to study propagation and DX. At present he has a FRG7 fed by a pair of home-brew 10 m crossed dipoles and is constructing a receiver for 2 m plus a general purpose VHF/UHF receiver using varicap tuners into an IF strip. During the early mornings of July 1st and 3rd John received 539 signals from the Cyprus beacon 5B4CY and at 1300 on June 25th (that day again) GB3SX suddenly came up from a quiet 10 m band giving a 599 signal in Scotland. Like myself, and Nigel Golds BRS 36910, West-Chiltington, Sx, John has kept an ear open for the German beacon DLOIGI $(28 \cdot 195 \mathrm{MHz})$ and often receives it at good strength. On 16 of the 27 days between June 21 and July 18 I received DLOIGI at different times and at signal strengths varying between 539 and 599.
At 1940 on July 5 Nigel heard ZP5 calling EA on 10 m and around the same time on the 18th he heard UP2 and SM working among other European stations during one of those short skip openings which are becoming a regular feature on 10 m . John Branegan also sent an Oscar report. He is delighted to find that on some days he can hear 10 consecutive orbits of Oscar-7 and his log for July 2nd shows that he has heard signals from EA, DB, F, G, GM, OE, OZ, PA, I3, W1 and W3 coming from the satellite. At 1604 on June 27, Ian Jones G8LWI Old Heathfield, Sussex, heard a strong burst of signal on 2 m SSB from EA, but later that day there was more excitement when
he had a solid copy contact with G8LGQ, 15 miles away, while running 15 milliwatts to a 4 element quad!

Around 0600 on July 9th the atmospheric pressure started to rise rapidly and continued to do so through the 10th and then began to fall at midday on the 11th and, as usual, Alan Baker, recognising the ingredients for a tropo-opening, kept at his rig from 1900 on the 10th until 0448 on the 11th, during which time he worked ON, PA, F, SM and DJ. The first surprise of this opening came for Alan, and maybe others, at 2014 when he worked GM3JNW, 59, via the Belgian repeater on R4, ONOOV. Then at 2105 he worked G8JNJ Manchester via another Belgian repeater on R8, ONOHT. His best DX was at 0120 when he contacted SM6GUS on 2 m SSB at 1035 km .

At 2207 on the 10 th I received a 53 signal from GM30LK/P Peterhead, and for most of the 10th and 11th I could hear the 70 cm beacon GB3SUT from Sutton Coldfield with only a dipole feeding the receiver. That same dipole proved that 70 cms was well used during the VHF NFD on July $2 / 3$ because, between two separate one-hour sessions I heard 30 stations from Dover to Dorchester, and Arundel to Ipswich. Many of these stations were also arranging for contest skeds on 23 cm , which leads us into microwaves and another interesting report from Peter Kerry G8ARO Farnham, who has worked from virtually all the "high spots" on the South Downs, in co-operation with G8BDJ, G8GKV and G3JHM. Using 0.9 W ERP, a small klystron to a 14 dB horn, some paths, 35 to 60 km , have been worked many times. "However," says Peter, "one of the biggest events for me was the 10 GHz cumulative on June 19th", I can imagine the excitement, when he worked G3KSU/P on St Catherines Hill, IOW, a distance of 85 km from a good site 11 km SW of Newbury at $59+$, and later he received a 58 report from GW4BRS/P on Mynydd Maen, near Pontypool, over a distance of 116 km on a path obstructed in three places.

Both G8ARO and G8BCO took their microwave gear to the Farnborough and District Radio Society NFD site at Farnham Heights and at 0330 on July 3rd G8ARO detected the signal from the London 3 cm beacon GB3LBH near Romford, using a hand portable receiver consisting of an intruder alarm into a pocket FM radio at 108 MHz and a horn aerial with approximately 6dB gain! This prompted G8BCO to drive home and collect a larger rig and they both monitored this signal, the first over a 75 km path, often at $59+$ until it faded out around 0700. Summing up in his letter, Peter said, "Must have been one of the best all-time land ducts". Well done, both of you, and we look forward to hearing more in the future.

On June 24th Cmdr Henry Hatfield, Sevenoaks, observed a plage and a solar flare with his spectrohelioscope and recorded several bursts of noise on his 136 MHz radio telescope. This was the beginning of a period of solar activity which lasted until July 4. Both Henry and myself recorded many individual bursts and some longer periods of solar noise at 136 MHz throughout the period. This event was confirmed by John Smith, Cranleigh, with his 142 MHz telescope and by Cliff Ranft using his cosmic noise equipment at 30 MHz .

Once again we have covered amateur activity from 10 m through to 3 cm and natural disturbances occurring on the sun 93 million miles away and in our atmosphere, relatively a few miles above us. Thank you all for your reports and letters and for the interest you are taking in my column.

DESIGN YOUR OWN PROJECTS-2-continued from page 438 about be OK. This is running the BFY51 pretty close to its limits but even Toby should find it difficult to achieve this rather exceptional "half shorted out" condition.

With a complete short circuit the power dissipated in Tr 2 will be much lower since Vce will only be about half a volt. An unfortunate side effect in this state is that Rl will now be dissipating over 3 W so we are going to have to replace this with a 5 W type. The dissipation in $\operatorname{Tr} 1$ when shorted is going to be around 18 W which is pushing things a bit! However, it should be pointed out that the protection circuit is meant to handle safely the occasional accident and should not be subjected to studied abuse.

Well, the unit worked as built and is still helping to entertain Toby on long journeys.

## Postscript

We would consider this circuit rather dubious if it were to be used as a test-bench supply for general experimenting. In this type of environment a supply ought to be able to cope safely with all forms of overload conditions for quite long periods. At least the supplies we use have to possess these characteristics if they are to survive! However, we feel that the simple short-circuit protection we have used is all that is required in this application.

Next month our authors will deal with another little problem found around a car. The courtesy light that goes out just at the wrong moment!

## HInDLV IOTE:



SO YOU WANT TO PASESTHCHAE? - 1 September 1824
The isfaB has intomed us Hiat lince the athore article went to pitss the prices of theli puthications mentionid in the artiele laye been increased The "A suide to Amaterr Radio" is now £1. 38, "She Radio Amatcurs' Examination Manual is $\mathrm{f1} .60$ and the "Radio Amateurs Examination Retision Notes" is $86 p$. These prices are inclusive of postage.

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| AC142K | 0.25 | BA156 | 0.13 | BC214 | 0.17** |
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| ACY18 | 0.65 | BC108 | 0.12 | BC307 | 0.20** |
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| AD161 | 0.75 | BC117 | 0.22* | BCY30 | $1-00$ |
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| AF106 | 0.45 | BC125 | 0-18** | BCY32 | 1.00 |
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## BREAKDONN PARCER-

four unused made for computer units containing most useful components, and these components unlike those from most computer panels, have for instance have leads over $1^{*}$ iong the diodes have approx. $\frac{1}{2}$ " leads. List of the major components is as follows: 17 assorted transistors, 38 assorted diodes, 60 assorted resistors and con densers, 4 gold plated plugs in units which can serve as multipin plugs or as hook up boards for experimental or quickly changed circuits (note we can supply the socket boards which were made to receive these units). The price of this four units parcel is f I including VAT and post (considerably less than value of the transistor or diodes alone). Don't miss this splendid offer.

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## IT'S FREE

Our monthly Advance Advertising Bargains List gives details of bargains arriving or just arrived-often bargains which sell out beiore our advertisement can appear ing list and it's free-just send S.A.E. Below are a few of the Bargains still available from previous lines.
SPECIAL NOTES: The "+ +" sign after the amount shows the amount of VAT. The postage item shown is based upon the amount an article costs to send if this only a few items are ordered her parcel. Where one or the minimum parcel postal charge of $50 \mathrm{p}+$ YAT and the VAT rate would be $8 \%$ or $12 \frac{1}{2} \%$ depending upon whether the article ordered was rated at $8 \%$ or $12 \frac{1}{2} \%$ that is the Customs and Excise rule).
Nearly Sold out. Car Starter Charger kits-we have been able to get a few more of the rectifiers which made this kit possible at a very low price but had to
pay more for these and with the increased postal charges, price of this is now $\mathbf{8 7} \cdot 95$. It is still a bargain, however, and it is interesting to note the various uses to which our customers have put this kit. One wrote in to say that he started his old Bentley with it, apparently it was almost impossible to turn over by hand but started quite quickly with our car starter. Another customer wrtes to say that fitted on to his electric lawn mower. the battery of which had worn out, he now uses the car starter to drive the mower found for our various kits and welcome hints and suggestions from customers.
Automatic Telephone Exchange, this takes standard GPO instruments which can dial each other, up to 75 telephones can be interconnected. Believed to be in good working order in fact
removed recently from a Bank by the builders doing alterations. The exchange which is floor standing is full of relays and uniselector switches and has a separate power units supply for the 50 v AC bells and the DC for speech. Price of this exchange is $£ 250$, carriage at cost, telephones are not included in this price but are available. Prices $£ 3+24$ p or new style $£ 5+40 \mathrm{p}$.
Tubes for Rigonda 6" TV's. Limited quantity of these are available, used but tested and guaranteed o.k. $1 \cdot 50+18 \mathrm{p}$.

Power Units for Rigonda 6* T.V. Again not new but tested and guaranteed. Price $\mathbf{\varepsilon 3 \cdot 5 0}+44 \mathrm{p}$. Post $40 \mathrm{p}+4 \mathrm{p}$.
Fan Motor, mains operated, this is totally enclosed and therefore suitable for extracting dusty or corrosive $1300 \mathrm{rpm}, 240 \mathrm{v} ; 50 \mathrm{hz}, 7 \mathrm{w}$. Spindle is threaded to take the fan blade, no doubt could be adapted. $£ 1 \cdot 50+$ 12 p . Post $60 \mathrm{p}+5 \mathrm{p}$.
12v Battery Motor, Delco, as used for blower heaters, fans etc. This is very powerful but quite compact. size spindle. this is a series wound motor so it will also work off AC and can be made reversible by bringing out the internal brush connections to a d.p. changeover switch. Price $22 \cdot 00+16 \mathrm{p}$. Post $50 \mathrm{p}+4 \mathrm{p}$.
"C" Core Transformer, primary tapped $115 \mathrm{v}, 200 \mathrm{v}, 240 \mathrm{v}$, primary screen to separate tag and 4 secondaries. (1) is $50-0-50 \mathrm{v}$ @. 9 A (2) $17 \mathrm{volts} \cdot 7 \mathrm{~A}$ (3) 17 volts. 7 A
(4) 20 volts @ GA . It will be seen that by interconnecting it could be made to give $50 \mathrm{v}-0-50 \mathrm{v}$ at 900 mA , a useful transformer for high power ampintiers etc. Ex equipment but guaranteed perfect $23.75+28 \mathrm{p}$. Post $£ 1 \cdot 00+8 \mathrm{p}$ The makers price of this is over $£ 10$
Professional Scotch Tape on 10t" spool (these having the normal $\pm^{*}$ spindle). We understand that this spool is standard for most popular professional reel to reel tape recorders. This is frst class tape normanty priced new and unused Price $84.50+36 \mathrm{p}$. Post $50 \mathrm{p}+5 \mathrm{p}$. new and unused. Price $\mathbf{4} \cdot \mathbf{5 0}+36 \mathrm{p}$. Post $50 \mathrm{p}+5 \mathrm{p}$. Ex G.P.O. Telephones. We have recently had to replace our stocks of these and like everything else the prices are up so we take this opportunu or remin. our prices. Three types are avainable-standard desk moder, this is he prie dial but no interal bell price $£ 2.00+16$ p post $80 \mathrm{D}+70$ Model 3 has no dial but internal bell price $\mathbf{~} 2 \cdot 00+\mathrm{i} 6 \mathrm{p}$. Post $\mathrm{f1} \cdot 20 \mathrm{p}+9 \mathrm{p}$.
Sundries available. 50 v transformer for ringing GPO type bells. price $22 \cdot 00+16 \mathrm{p}$. Post $40 \mathrm{p}+3 \mathrm{p}$. Twin connecting wire for teiephones. Daketre con, price
$\mathbf{5 5} \cdot 00+40 \mathrm{p}$. Post $80 \mathrm{p}+7 \mathrm{p}$. Bakelite cased bells, so you can hear telephone when you are not in same so you can hear telephone when you are
room, price $£ 2 \cdot 50+20 \mathrm{p}$. Post $50 \mathrm{p}+4 \mathrm{p}$.
Kymograph Brodie Starling. motor gear box type. This is a mains operated unit very solidly constructed in heavy cast iron case. It seems to be bascaty a motor quoted in mm per minute, on 9 * diameter cylinder but the drive which is fitted to the device is normal $\frac{10}{}$ spindle and the speeds are selected by a knob on the front dial through which the knob rotates. is calibrated as follows. $2,4,8,16,32,64,128,256,512,1024$. We are not at all sure to what purpose these machines were normally put and would welcome any information about them from readers. We have only a few, price $\mathbf{8 1 7} \cdot 50+£ 1 \cdot 40 \mathrm{p}$. Post $£ 1 \cdot 60+14 \mathrm{p}$
Interrupted Beam Switch Kit. This has been recently re-arranged and is suitable for operation by a normal light beam or an infra red beam. The kill consists photo electric cell, 2 transistors, relay and all the necessary resistors and condensors together with mounting board and tag strip. This is both useful and educational, price ${ }_{\mathrm{E} 2} \cdot 00+16 \mathrm{p}$. Post $50 \mathrm{p}+4 \mathrm{p}$.


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I bought this beautiful little milliamp-meter through the special Bargain List that Home Radio sent me when I bought a copy of their famous Components Catalogue. I reckon I saved a good $£ 1.50$ on this one item alone-more than the catalogue itself cost. If I'm not mistaken, you got

that natty transformer the same way, so no doubt you're as familiar as I am with the scores of bargains they're offering. But have you noticed that there are quite a few bargains in the main catalogue itself? That Resistor Pack SP22 for in-stance-400 preferred value $\frac{1}{4}$ watt $5 \%$ resistors for only £6, beautifully arranged for easy selection. If that's not a bargain, what is?
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# VAVEAMPLIFIER Build Chris Rogers' SIRAC MK 1 stereo design 

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TEN CASSETTE DECKS COMPARED
Gordon King and Fred Judd join forces in another big cassette deck comparison with reliability and performance checks - and a comprehensive 'panel listening comparison'.

## BUDGET HI-FI SUPPLEMENT

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## five yamaha cassette decks to be won

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## 30

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$\begin{aligned} & \text { 8p) } 220-8 \mathrm{p} \text {. (50V-40p) } 470-11 \mathrm{p} \text {. (50 } \\ & 15 \mathrm{p} .100 / 25 \mathrm{~V}-48 \mathrm{p} .1000 / 50 \mathrm{~V}-22 \mathrm{p} \text {. }\end{aligned}$
$\begin{aligned} & \text { Subminiature bead tantalum electrolytics }-0.1,0.22, \\ & 0.47,1-0,2-2 @ 35 \mathrm{~V}, 4 \cdot 7 / 25 \mathrm{~V}-11 \mathrm{p}, 10 / 25 \mathrm{~V}-43 \mathrm{p} .20 / 16 \mathrm{~V}\end{aligned}$
$\begin{aligned} & 0.47,1 \cdot 0,2 \cdot 2 @ 35 \mathrm{~V}, 4 \cdot 7 / 25 \mathrm{~V}-11 \mathrm{p} .10 / 25 \mathrm{~V}-43 \mathrm{p} .22 / 16 \mathrm{~V} \text {, } \\ & 47 / 6 \mathrm{~V}, \& 100 / 3 \mathrm{~V}-15 \mathrm{p} .\end{aligned}$
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$47 \mathrm{pf}-3 \mathrm{p}$. 56 pf . to 330 pf .-4p.
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Mull
$\begin{aligned} & \text { Mullard polyester } 250 \mathrm{~V} \text { vertical mounting E6 series. } \\ & 0.01 \text { to } 0.1-4 \mathrm{p} .0 \cdot 15,0.22-5 \mathrm{p} .0 .33,0 \cdot 47-8 \mathrm{p} .0 .68-11 \mathrm{p} .\end{aligned}$
$\begin{aligned} & 0.01 \text { to } 0.1-4 \mathrm{p} .0 .15,0.22 \text {-5p. } 0.33,0.47-8 \mathrm{p} .0 .68-11 \mathrm{p} . \\ & 1.0-13 \mathrm{p} .1 \cdot 5-20 \mathrm{p} .2 .2-22 \mathrm{p} .\end{aligned}$
Mylar (Polyester) Film $\mathbf{4 0 0 \mathrm { V }}$
$\begin{aligned} & \text { Mylar (Polyester) Film 100V vertical mounting. } 0.001 \text {, } \\ & 0.002,0.005-3 \mathrm{p} .0 .01,0.02-4 \mathrm{p} .0 .04,0.05-4 \mathrm{p} .\end{aligned}$
$0.002,0.005-3 \mathrm{p} .0 .01,0.02-4 \mathrm{p} .0 .04,0.05-41 \mathrm{p}$.
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[^3]
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