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VOL. 53 NO. 1 ISSUE 843 May 1977

BRITAIN'S PREMIER MAGAZINE FOR THE DO-IT-YOURSELF RADIO AND ELECTRONICS CONSTRUCTOR

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FREE THIS MONTH!

THE SECOND DOUBLE SIDED INFO-CARD— Semiconductor Characteristics.

ALSO:

4 page INDEX listing all articles, letters, Production Lines, 'Kindly Note', etc., for Volume 52, May 1976 to April 1977.

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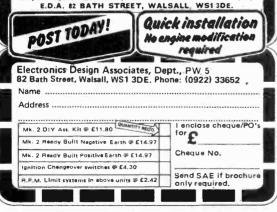
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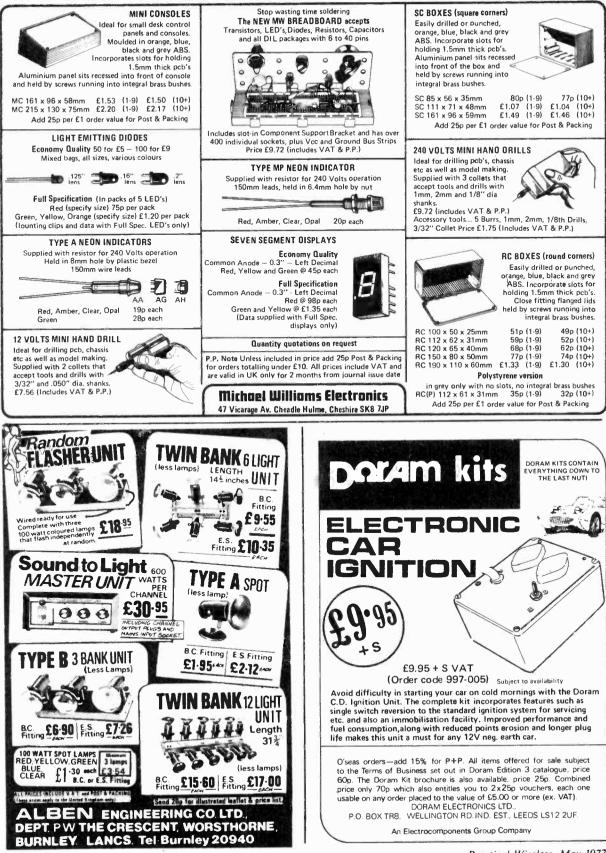
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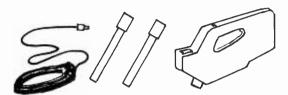




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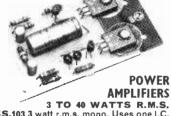
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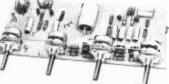
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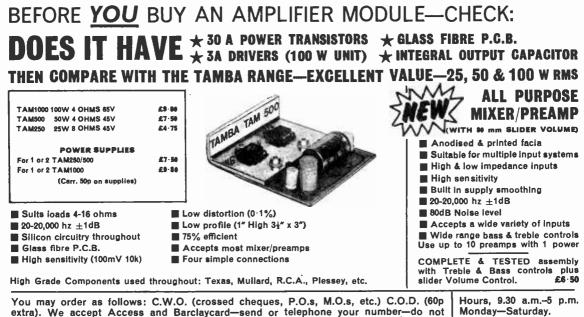
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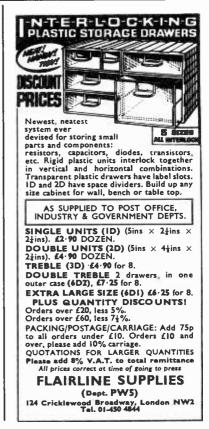
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1A7GT 60 6BE6 40 6L1 2 50 12J5GT 40 85Z3 80 DK96 70	ECL84 65 E281 40 PCL82 40 UBC81 55 2759 5.85 AF121 35 0A81 11 ECL85 70 GY501 95 PCL83 49 UBF80 50 Transistors AF124 . 38 0A85 11
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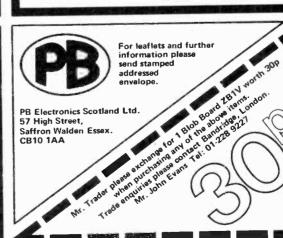
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PCB ETCH KIT 3 ITEMS 22* ETCH RESIST PEN 2TIPS 75p* FEC ETCH PAK TUB 800gm f1* 6x4" SRBP 45p*NYLON F/Gf1*

SCR AND TRIACS BR100 25p* TAG 1A400V 50p*1A600V 69p* 1A50V 37p*,C106D 4A400 60p* TRIACS:SC146D 10A400V f1*

DISCO TRIAC 15A400V £2*

AUDÍBLE WARNING BLEEPER 12V35mA £1,20*10 off £1* CAPACITORS 22pf - ,o1 5p ELECTROLYTIC IN 10 & 25V 1/2/10/50/100 7p 50V 10p 200/500 10p.1000/25 20p

POTENTIOMETERS AB etc20p PRESETS 6P 4/RESISTORS 2p HEATSINKS T05 or 18 7p T03 16p,T03 4"finned 50p DIN:PLUGS all 15p,Sock 10p SWITCHES SPST 20p Dpdt 29p

GAS DETECTOR TGS 308etc:4*

VERO 0.1" PITCH COPPERCLAD 24"x5" 40p* 34"x5" 45p* 24"x34"36p* 34"x32" 40p* 31x17" f2 PACE CUTTER 65p* DLL BREADBOARD 2"x4" f1 or 6"x4" f2,VERO PINS 36 30p*

BLACK PLASTIC CASES 42mmx.

80x60 50p*100x75 60p*99x 120 70p,DESOLDER BRAID 50p

vero

FAST SERVICE ON EX STOCK ITEMS



SEVEN SEGMENT LED DISPLAYS BRIGHT DL707 COM ANODE & DL704 COM CATH, 0,3" f1* DL747 0.6"JUMBO CA f1,25* DIGITAL CLOCK IC 51224 £4* ZENON STROBE TUBES £4 & £7*

RED LEDS IOP.

RED LEDS 209 STYLE 0.125" OR 0.2" DIA. NO CLIP 10p*

TIL209 RED LED & CLIP 12p* BIG 0.2" RED LED & " 14p* GREEN OR ORANGE LEDS 29p* ORP12 54p*2N5777/OCP71 34p*

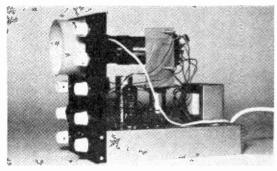


INDUSTRIAL, EDUCATIONAL, TRAD & EXPORT SUPPLIED, SEND FOR OUR FREE CATALOGUE LIST SAE EDUCATIONAL, TRADE MANUFACTURERS EXCESS STOCKS PURCHASED DISCOUNTS 10% OFF 100 up. 15% OFF 1000 up.,.



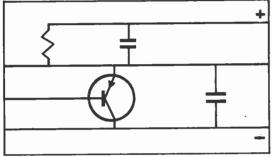
556 2 x 555 90,* LM381 Preamp 12 ArTCHING 22p Pr* TIP31632668 703 RF/1F AWE 29b LM3800 4x0PA 53p* AC127/R 176 15p* TIP41/42e 161, 26 FT 710 DIL14 35p* MC1310 stereot2 BC107 Bp* TIP31632668 FT 712 Regulator 44p* MC1310 stereot2 BC107 Bp* TIP3055 65p* 714 DIL 4pin 31p* MC1330 75p BC108 Bp* TIP3055 65p* 741 DIL 4pin 31p* MC1330 75p BC108 Bp* TIP3055 65p* 744 DUAL 741 69p MPC6030 11 BC108 Bor C 13p* IN4001/2 4p* FT 745 DIL 8 pin 32p* NES56 PTOPA 12* BC147/8/9 Bp IN4001/2 4p* FT 745 DIL 6 pin 32p* NES56 PTOPA 12* BC147/8/9 Bp IN4007 7p* FT 7805 S71A*reg11.5* NE560 cr 55 90p* BC147/8/9 Bp IN2007 7p* FT S017/4/9/102 IN2026 ory 7p* 7031 6F AF 11.5 NE560 cr 55 90p* BC177/8/9 12p ZN2064 UT 40p* S07601 14*			and the second se	
555 TIMER 340* LM380(60748)*11 INS DISE SET 50* ed.* TIP31/32.500* 555 2 F 55 20 900* LM380(60748)*11 INS DISE SET 50* ed.* TIP31/32.500* 703 TR07 ALD 1250* MC1300 AXDA 530* MATCHINO 200 F** TIP31/32.500* 703 TR07 ALD 1250* MC1300 AXDA 530* MATCHINO 200 F** TIP31/32.500* 703 TR07 ALD 1250* MC1300 AXDA 530* AL127/16 176 150* TIP41/428* E656 710 DILL4 Spremarcia MC1300 AXDA 530* AL127/16 176 TIP41024*** E566 723 Regulator 4400 Spremarcia MC1300 SATA Tip TIP3056 E597 741 DIL 4pin 310* WC1300 SATA Tip TIB413 UT 260 TIP41/428** LM380/172 741 DUL 741 699 339* WC40008 H* 729 EC1080 or C 129* TIN4012/2 447 740 DUL 741 690* MAS 01.50 SSS 500 500 SS EC147/8/9 E9 2N306/8 149* 7805 SUA+reg(1.155* NS560 FT 749 EC147/8/9 E013/3/4d/110 2N306/9 170* 7812 A 7815eet1.40* NS55 TIMER 349* EC187/6/9 EC187/6/9 2N306/9	⇒	TOP DISC	COUNTS .	NEW LOW PRICES.
555 TMER 34P* LM380(40748) ft IFS BUSB BET 5pect 5pect TID31/32.500F 705 T09 T09 or DIL290 LM3800 4x0P AS3p* AC127/8 176 15p* TID31/32.500F TID31/32.500F 705 T09 or DIL290 M1300 4x0P AS3p* AC127/8 176 15p* TID31/32.500F TID31/32.500F 705 T09 or DIL290 M1300 4x0P AS3p* AC127/8 176 15p* TID31/42.508p TID41/4 486 155 728 Regulator 44/P MC1310 stereoc2 Bp TID31/42.505 Sp* TID31/42.505 741 DIL 14pin 31p* MC1319 Preampic 2 BC107 Bp* TID34/3 UIT 26p Sp 741 DUL 14pin 31p* MC1319 Preampic 2 BC108 Bp* TID34/3 UIT 26p Sp 741 DUL 741 69p Bp* MC6000 ft Pt BC108 0r C 13p* IN01/4146 4p 741 DUL 741 69p* MC5050 1r EC147/8/9 Bp IN000/7 7p* 7805 S701A+reg(1.35)* MS540 or 550 fc2* BC167/8/9 112 2x706/8 140* Sp 7400 140* MT555 TMER 34p* BC127/8/9 16p 2x0304/5pp28p* Ar51224 CLOCK 14 * NT556 C11F ft BC137/4/2110p 2x0306 y 77* 2x0305 115m45p* 2x030/5/11 5p* 2x03		INTEGRATED CIRCUITS		TRANSISTORS AND DIODES
741 DIL Bpin 21p* WC1330 75p BC108 Bp* Tis33 UT 26p 741 DIL 14pin 31p* WC1330 75p BC108 DC 13p* IN914/4148 4p 741 DIL 14pin 31p* WC3300 Tis39 Presmpt2 BC108 DC 13p* IN914/4148 4p 741 DUL 741 699 33p* WFC40008 iW 72p BC108 DC 13p* IN914/4148 4p 748 DIL 8 pin 32p* WFS36 FETOPA 12* BC109 Dp* IN4001/2 5p* 7805 SV1A+reg(1.35)* VFS36 Or 550 C12* BC167/8/9 Bp 2x706/8 149* 76103 KW AF 61.50 NE556 TIMER 34p* BC167/8/9 12p 2x8064 UT 40p* 2x706/8 14p* 7603 Stgen 14 NE560/1/2/5/6/73 BC187/8/9 12p 2x8064 SUT 405 17p* CL8038 Stgen 14 SN76660 IF 74p BC137/3/46/1122ea 36p* 2x8055 90W 33p* 2x8055 90W 33p* 2x805/90W 33p* 2x805/6/7 9p 2x805/6/7 9p 2x805/6/7 9p 2x8070/2/3/10p 2x8070/2/3/10p 2x8070/2/3/10p 2x8070/2/3/10p 2x8319 A22e17p 2x804/5/6/7 5p* 2x8070/2/3/10p 2x8070/2/3/10p 2x8319 A22e17p 2x805/6/7 p		556 2 x 555 90p* 703 RF/IF AMP 29p 709 T099 or DIL26p* 710 DIL14 35p*	LM381 Preamp £2 LM3900 4xOPA 53p* MC1303 2xPre £2 MC1310 stereo£2	INS BUSH SET Sp & a* TIP31/32.50p* MATCHING 20p Pr* TIP31C32c68p* AC127/8 176 15p* TIP3142em 66p* AD161/162 em36p* TIP41c42c"f1.50* BC107 8p* TIP2155 68p*
741 DTL 14ph 31p* MC1339 Presuppi2 EC108B or C 13p* IN0401/2 4p* 741 DTL 14ph 31p* MC1339 Presuppi2 EC108B or C 12p* IN4001/2 4p* 747 DUAL 741 69p* MCC6030 11 EC108B or C 12p* IN4001/2 4p* 747 DUAL 741 69p* MS536 PETOPA L2* EC107/8/9 Bp IN4007 7p* 7805 571A+regc1.35* MS540 or 550 C2* EC157/8/9 1p 2N706/8 1q* 7815a81.40* MS556 2x555 90p* EC177/8/9 Bp 2N706/8 1q* 7613 6# AF C1.50 MS556 2x555 90p* EC177/8/9 16p 2N306/07/7p* 2N306/07/7p* CA3046 54p* SN76711 Fs 11p EC12/3/4a/112p 2N3055 117* IC16038 Sigen f4 NS7660 1F 74p EC137/8/9 2N3055 115#45p* LM300H f1 * SN7660 1F f4p EC137/8/2 12p* 2N3055 115#45p* 2N3055 115#45p* LM300H f1 * SN7660 1F f4p EC137/8/2 12p* 2N3057 115#45p* 2N3072/3/4 100 15# 2N3072/3/4 100 15# 2N3072/3/4 100 15# 2N3072/3/4 100 15# 2N30702/3/4 100 15#				
741 T099 33p* HFC4000B H* 72p Dilos 1p* 1N4004 5p* 744 DULA 741 699 HFC4000B H* 72p Dilos 1p* 1N4004 5p* 746 DULA 741 699 HFC4000B H* 72p Dilos 1p* 1N4004 5p* 748 DIL 8 pin 32p* NE536 FETOPA 12* BC137/8/9 Hp 2N706/8 1p* 7805 SV1A+reg(1.3) NE555 TIMER 34p* BC167/8/9 1p 2N706/8 1p* 7613 6% AF 61.150 NE556 Zx555 90P BC137/8/9 1p 2N3063 170* CA3046 54p* SN76660 1F 74p BC137/8/9 1p 2N3053 170* LM300H 1S3p* TBA800 5% AF 84p BC137/3/4a/112e 2N3053 170* LM301K SV reg (2: TBA810 7% AF 84p BY50/51 15p* 2N305/6/7 9p 2N3065/15w45p* 2N306/6/7 9p LW308H T05 can (1* ZM14 Radio (1,19 BSX 19/20/21/3p* 2N306/6/7 9p 2N306/6/7 9p 2N306/6/7 9p 2N306/6/7 9p 2N306/6/7 9p 2N306/6/7 9p <td< th=""><th></th><th></th><th></th><th></th></td<>				
747 DUAL 741 69.* HEC6030 11 EC109B or C 128.* INA007 77 748 DIL 8 pin 329.* NES36 PETOPA L2* EC137/8/9 Bip INA007 77 7805 SVIA+reg(1.35* NES40 or 550 C2* EC137/8/9 Bip INA007 77 7815.861.10* NES55 STHER 34.* EC157/8/9 Ip 2N3064 U14 of 7613 6# AF C1.50 NES56 2x555 90.* EC177/8/9 Isp 2N3065 U14 of AK51224 CLOCK (4 NES56 017 74p EC12/3/44/112p 2N3059 179 LA300H E1 * SN76611 IF F1 EC12/3/44/112p 2N3059 179 LM300H C1 * SN76611 IF F1 ED131/132ea 369.* 2N3051 I15#459* LM300H C5 × reg (2 * TA414 Radio (1,19 EN157/51 EN16/20/11 159* 2N307/6/7 BF LM300H C5 × reg (1 * 7490 39* MEE265 S9* 2N308/9 A22179 MI22055 BF 2N307/6/7 BF 2N308/9 A22179 MI22055 S9* 2N308/9 A221470 179* <th></th> <th></th> <th></th> <th></th>				
748 DIL 8 pin 32p* NES36 FETOPA 12* BC137/8/9 Bp TN0507 Tp 7805 SV1A+reg(1.35) NES36 Or 550 12* BC137/8/9 12 ZN706/8 14p* 7812 A 7815eat1.40* NES55 TIMER 34p* BC167/8/9 12 ZN706/8 14p* 7805 SV1A+reg(1.35) NES56 ZX555 909 BC177/8/9 16 ZN706/8 14p* AT51224 CLOCK 14 * NE566/1/2/5/6/73 BC187/8/9 12 ZN306/8 ZN306/8 17p* CA3046 54p* SN76660 1F 74p BC177/1/2 15p* ZN3053 17p* ZN3053 17p* LM300H 61 SN76660 1F 74p BC170/1/2 15p* ZN3053 17p* ZN3065 115w45p* LM308K 5V reg t2* TBA800 5% AF 84p BY50/51 15p* ZN306/6/7 9p LW308H 705 can 61* ZN414 Radio 11 BSX 19/20/2113p* ZN306/6/7 9p ZN306/6/7 9p LW308H T05 can 61* ZN414 Radio 61, 74p T5p* C160D SCR S4p* ZN306/6/7 9p ZN306/6/7 9p ZN306/6/7 9p ZN306/6/7 9p ZN306/6/7 9p ZN306/				
7805 571A+reg(1.35* YB540 or 550 (2*) BC157/8/9 (1) 2x70364 U14 409* 7812 at 1.40* NE555 TIMER 349* BC157/8/9 (1) 2x70364 U14 409* 76013 GW AF (1.50) NE556 2x555 900* BC177/8/9 (16) 2x70364 U14 409* AT51224 CLOCK t4 NE556 (1/2/5/6/718) BC187/3/46/112p 2x1926 or 7 CA3046 54p* SX7660 IF 74p BC187/3/46/112p 2x1936 or 339* LM300H f1 * SX7660 IF 74p BC187/3/46/112p 2x1936 or 339* LM301A D11 8 33p* TA860 5W AF 84p BC137/3/46/112p 2x1935 501 339* LM309H T05 can (1)* ZX14 Radio C1 1 BC187/3/20/2112p* 2x19705/6/7 9p LM309H T05 can (1)* ZX14 Radio C1 1 BCX76/51 15p* 2x13708/9 9p LM309H T05 can (1)* ZX14 Radio C1 1 BCX82 ZENER 10p 2x13708/9 9p LM318 70Vus0PA (3*) NEW 1466/69 64 BEX88 ZENER 10p 2x13708/9 9p LM309 H T40* T400 12p* 7490 39p* ME3055 65p* 2x1830 AE17p 4000 14p* 7400 12p* 7490 39p* MUE3055 65p* 2x1830 AE17p 4000 14p* 7401 15p* 7401 15p* T182/93 400 T182/93 400 EX1607 EX132p* 4000 1		748 DIL 8 pin 32p*		
7812 & 7815 ext1.40* NE555 TIMER 34p* BC167/8/9 12p 228364 Urt 40p* 76013 & WA F C1.50 NE555 CX555 90p* BC177/8/9 15p 20204/5pnp28p* AT51224 CLOCK 14 * NE560/1/2/5/6/73 BC182/3/4a/110p 20204/5pnp28p* CA3046 54p* SN76660 IF 74p BC127/3/4a/112p 201053 17p* LM300H 1 SN76660 IF 74p BC127/3/4a/112p 201053 17p* LM300H 61 SN76660 IF 74p BC170/1/2 15p* 201053 17p* LM300H 61 SN76660 IF F4p BC170/1/2 15p* 2013702/3/4 103 15p* LM30BK 5V reg (22 TBA800 5% AF 84p BY50/51 15p* 20105/51 15p* 20105/6/7 9p LM30BK 70Vus0PA (3* NEW 1466/69 (4 MF28055 9p* 203706/6 9p 203706/6 7p 9p 203819 & 821e17 40051/11 15p* 749/59 74p* 74101 7400 <td< td=""><th></th><td>7805 5V1A+reg£1.35*</td><td>NE540 or 550 £2*</td><td></td></td<>		7805 5V1A+reg£1.35*	NE540 or 550 £2*	
Ar51224 CLOCK 14 * NE560/1/2/5/6/73 CA3046 540* SN72741 as 741 ICL8038 Sigen 14 * SN76660 IF 74p LM300H f1 * SN76660 IF 74p LM300H f1 * SN76611 IF f1 LM301A D11 8 33p* TA8800 5% AF 84p LM308 Sigen 14 * SN76660 IF 74p LM308 Sigen 14 * SN76660 IF 74p LM308 Sigen 14 * SN76660 IF 74p LM308 Sigen 14 * SN76611 IF f1 LM308 Sigen 14 * SN76610 7% AF 94p LM308 Sigen 14 * SN76610 7% AF 94p LM308 Sigen 14 * SN76610 7% AF 94p LM308 Sigen 14 * SN76660 IF 74p LM308 Sigen 14 * SN76660 IF 74p LM308 Sigen 14 * SN76670 F% AF 84p LM308 Sigen 14 * SN76670 5% AF 84p LM308 TO% LAB * SN76676 7% AF 84p LM308 TO% LAB * SN8166/69 f4 * CMOS 5 TFL * SN1166/69 f4 4000 14p* 7400 12p* 7400 10p* 7401 12p* 7490 75p* T1228 SENER 10p 2N3706/6 79 2005 56 50p* 2003 2007 16p* 7402 15p* 7482/93 5703 99p* 2N3819 A232e17p 4000 14p* 7401 15p* 7490 7495 75p* T1228 30 40p* BRIDGELAS0 20p* 4007 16p* 7404 16p* 74106 12p* 7410 612* 4007 16p* 7404 16p* 7410 12p* 7410 732P* PAK A: 11 RED LEDS full spec 61* 4011 16p* 7411 71p* 74121 39p* PAK A: 11 RED LEDS full spec 61* 4049 50p* 7447 70p* 74141 77p* PAK Si 11 BCD LEDS full spec 61* 4059 20p* 7447 70p* 74141 74p* PAK 5; 71 BC 712 18* 7410 7412 28* 4051 26p* 7447 70p* 74141 74p* PAK 5; 71 BC 712 11 B2* 7416 174* 7415 1; 75p* PAK A: 11 RED LEDS full spec 61* 4051 26p* 7447 70p* 74141 74p* PAK 5; 71 BC 712 11 21% 2704 51 4551 26* 747772 27p* 74145 75p* PAK 6; 7 BYS1 1; 1 * H 2N3704 51 4511 45* 747774 31p* 74151 75p* PAK 6; 200 55 1; * H 25 C069 1; * 4001 140* 7457 645 755* PAK 6; 7 BYS1 1; * H 400114 1; * 50014 1; * 747374 51 75; * PAK 6; 7 BYS1 1; * H 40014 1; * 50014 1; * 747374 51 75; * PAK 6; 7 BYS1 1; * H 40014 1; * 74573 4; * 745373 4; * 745473 4; * 745373 4; * 7454373				BC167/8/9 12p 2N2646 UJT 40p*
CA3046 54p* SN72741 (as 741) 5212/3/42/112p 2N3055 907 33p* ICL8038 Sigen t4 SN76661 IF 74p EC12/3/42/112p 2N3055 907 33p* LM300H E1 SN76611 IF f1 EC13/3/42/112p 2N3055 907 33p* LM301A D11 8 33p* TA8600 5W A F8 4p EC13/2/3/42/112p 2N3055 901 33p* LM309K 5V reg t2 TTA810 7W AF 94p EKY53/551 15p* 2N305/511 sp* 2N305/677 9p LM309K T05 can f11* ZM14 Radio f1.1 BA11/20/2112p* 2N3708/9 9p 2N3708/9 9p LM309K T05 can f11* ZM14 Radio f1.1 BKS 11/20/2112p* 2N3708/9 9p 2N305/6/7 9p LM318 70Vus0PA f3* NEW 1466/69 f4 Imp 7490 Stress 50m Stress 2m304/5/6 15p VLE2055 50 9p* 2N3708/9 9p 4000 14p* 7400 12p* 7491 75p* TM29/93 672p* TM29/93 672p* NES304 A2m77p 4000 14p* 7401 15p* 7401 15p* 7410 f1: Stress 7410 50 50* ZM3904/5/6 15p DA86J1/91 50* Stress 208200 A2m77p 4000 15p* 7401 15p* 7401 15p* 7412 23p* PAM 55 57 50* Stress 2190 425/6 15p DA8500 A2m77p 4000 16p* 7402 15p* 7410 f1: Stress 7410 50* Stres 741050* Stres 74105 50* <t< td=""><th></th><td></td><td></td><td></td></t<>				
ICL8038 Sigen f4 * SN76660 IF 74p BCT70/1/2 15p* 2N3055 115w45p* LM300H f1 * SN76660 IF 74p BD131/132ca 36p* 2N3055 115w45p* LM301A D11 8 33p* TA8800 5% AF 84p BDT50/51 15p* 2N3055 115w45p* LM308H C5 v reg f2 * TA8410 7% AF 94p LM309H C5 can f1* ZN414 Radio 17 AF 94p BDT50/51 15p* 2N30576/7 3p LM308K T0% can f1* ZN414 Radio 11,19 BSX 19/20/21/3p* 2N305/6/7 9p CINCS TTL MUE2055 65p* 2N306/6/7 9p C106D ST TL MUE2055 65p* 2N309/6/7 9p 4000 14p* 7400 12p* 7400 13p* 7491 75p* 0081/91 MUE2055 65p* 2N3904/5/6 15p 4001 15p* 7401 13p* 7491 75p* 0081/91 MUE2055 65p* 2N3904/5/6 15p 4001 16p* 7400 16p* 7401 13p* 7497 33p* 74106 11.29* T1228 30 40p* BRIDGE1A50 20p* 4007 16p* 7404 16p* 740/6 72p* 74106 11.30* 57410 611.30* 5712 4011 16p* 7413 29p* 74107 32P* PAK A: 11 RED LEDS full spec fi* 4011 16p* 7413 29p* 7411 74p* 74123 69p* PAK A: 11 RED LEDS full spec fi* 4012 20p* 7444 72p* 74141 74p* PAK5 85p* PAK G: 7 BFYSI (1* H 82N3819 61* 4059 20p* 7447 72p* 74145 75p* 0PK A: 11 BCH2 1F 11 2N3704 fi 4538 (1* 7473/74 31p* 74151 75p* PAK B: 20055 (1* 4012 N3819 61* 4538 (1* 7473/74 31p* 74154 1.76* PAK P4K P4K5 755 0619				
LM300H 11 8 339 ⁵ TA860 5% AF 84 pt LM305K 5V reg t2 • LM305K 5V reg t2 • LM305K 5V reg t2 • LM309K 5V reg t2 · LM309K 5V reg t2 · LM300K 5V reg t2 · LM5005 ct = 1 12 · LM500K 50 · LM500S 50 · LM500K 50				
LANGORA D11 8 330* TARABGO 5% AP 84 p LANGORA D1 8 330* TARABGO 5% AP 84 p LANGORA 5V reg t2 * TARABGO 5% AP 84 p LANGORA 5V reg t2 * TARABGO 7% AP 94 p LANGORA 5V reg t2 * TARABGO 7% AP 94 p LANGORA 5V reg t2 * TARABGO 7% AP 94 p LANGORA 5V reg t2 * TARABGO 7% AP 94 p LANGORA 5V reg t2 * TARABGO 7% AP 94 p LANGORA 5V reg t2 * TARABGO 7% AP 94 p * 100 150* 7400 120* NET 1666/69 14 * 000 150* 7400 120* 7480 30* 20* 20* 20* 20* 20* 20* 20* 20* 20* 2				
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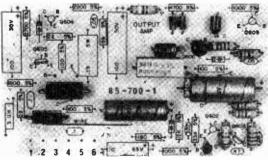
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	WB 5/7

High quality audio

OUR PRICE ONLY

Overall size 63mm. 13mm.

which should satisfy the most

The 450 Tuner provides instant programme selection at the touch of a button ensuring accurate tuning of 4 pre-selected stations, any of which may be altered as often as you choose, by simply changing the settings of the pre-set controls.

FP'

Used with your existing audio equipment or with the BI-KITS STEREO 30 or the MK60 Kit etc. Alternatively the PS12 can be used "If no suitable supply is available, together with the Transformer T461.

The S450 is supplied fully built, tested and aligned. The unit is easily installed using the simple Instructions supplied.

Max Heat Sink temp. 90C. • Frequency response 20Hz. Distortion better than 0.1 at 1kHz.
 Supply voltage 15-50v.
Thermal Feedback.
Latest Design Improvements. Load—3,4,5, or 16ohms. OSignal to noise ratio 80db.

Especially designed to a strict specification. Only the finest components have been used and the latest solidstate circuitry incorporated in this powerful little amplifier



FET Input Stage × VARI-CAP diode tuning *** Switched AFC Multi turn pre-sets LED Stereo Indicator

Typical Specification: Sensitivity 3µ volts Stereo separation 30db Supply required 20-30v at 90 Ma max.

STEREO PRE-AMPLIFIER

> A top quality stereo pre-amplifier and tone control unit. The six push-button selector switch provides a choice of inputs together with two really effective filters for high and low fre-

quencies, plus tape output.

PA 100

Frequency Response + 1dB 20H2-20KH2. Sensitivity of Inputs: 1. Tape Input 100mV Into 100K ohms 2. Radio Taner 100mV Into 100K ohms 3. Magnetic P. U. 3mV Into 50K ohms P. U. Input equalises to RIAA curve within 1dB ros 20H2 to 20KH2. Supply 20.5V at 20mA. Dimensiona-299mm × 89mm × 35mm.

£13.75 р & р 45р

MK60 AUDIO KIT: MK69 AUDIO KIT: Comprising: 2 x AL60's 1 x SPM60, 1 x BTM80, 1 x PA100, 1 front panel and knobs. 1 Kil of parts to include onioff switch, neon indicator, stereo headphone sockets plus instruction booklet. COMPLETE PRICE £29-55 plus 62p postage. Comprising: Teak veneered cabinet size 16¾" x 11¾" x 3¾", other parts include aluminium chassis, heatsink and front panel bracket plus back panel and approp-riate sockets etc. KIT PRICE £10-70 plus 62p postage.



7 + 7 WATTS R.M.S.

1 + 1 WATTS R.M.S. The Stereo 30 comprises a complete stereo pre-amplifier, power amplifiers and power supplr. This, with only the addition of a transformer or overwind will produce a high quality audio unit suitable for use with a wide range of inputs i.e. high quality audio unit suitable for use with a wide range of supplet of unstall all producing really first class results, the unit is supplied with rule range to producing really first class results, mains say supplied with rule results of your own construction or the cabinet stalled in a record plinth, rabinets of your own construction or the cabinet suplied with a minimum of installation of the cabinet fi performance with a minimum of installation difficulty (can be Installed in 30 mins).

TRANSFORMER	£2.45 plus 62p p & p	
TEAK CASE £5	•25 plus 62p p & p.	£16.2

critical A.F. enthusiast. ±4.35 ONLY Stabilised Power Supply Type SPM80 SPM80 is especially designed to power 2 of the AL60

25 Watts (RMS)

Amplifiers, up to 15 watts (r.m.s.) per channel simul-taneously. With the addition of the Mains Transformer BMT80, the unit will provide outputs of up to 1.5A at 35V. Size: 63mm, 105mm, 30mm. Incorporating short circuit protection.

INPUT VOLTAGE 33-40V. A.C. OUTPUT VOLTAGE OUTPUT CURRENT 33V. D.C. Nominal 10mA-1-5 amps £3.75 **OVERLEAD CURRENT** 1.7 amps approx. DIMENSIONS 105mm × 63mm × 30mm 80 £2.60 + 62p. postage TRANSFORMER

5 P & P 45p.

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THE MEDIUM POWERED 80

35[™]^{™S} power Amp!

ONLY

+ 8% VAT

A High Fidelity Power Amplifier with a maximum Power Output of 35 watt R.M.S., which has a maximum operating voltage of 60v. A MUST for all HI-FI users.

Maximum supply voltage Power output for 2% THD Harmonic distortion Load impedance Input impedance Frequency response +3dB Sensitivity for 25 watts O/P Max. Heat sink temperature Dimensions Mounting **Fuse requirements**

15-60v 35 watts R.M.S. 0.1% 3-8-16 ohm 50K ohm 20Hz-40KHz 280mV R.M.S. 90°C 102mm x 64mm x 15mm 2, 4BA fixing holes in heat sink

POSTAGE & PACKING Postage & Packing add 25p unless other-wise shown, Add extra for airmail, Min, £1.09

Enjoy the quality of a magnetic cartridge with your existing ceramic equipment using the new Bi-Pak M.P.A. 30 which is a high quality pre-amplifier enabling magnetic cartridges to be used where facilities exist for the use of ceramic cartridges only.

Used in conjunction are 4 low noise high gain silicon transistors. It is provided with a standard DIN input socket

for ease of connection. Supplied with full, easy-to-follow instructions.

AUDIO AMPL MODULES

The AL20 and AL30 units are similar in their appearance and in their general specification. Neweyer, careful selection of the plastic power devices has resulted in a range of uput powers from 5 to 10 watts R.M.S. The versatility of their design makes them ideal for use in record players, tape recorders, stereo amplifiers and caseste and cartridge tape players in the home. Harmonic Distortion Po = 3 watts t = 0.25% Load medance AL8 ohm Impedance 8-16 ohm

Frequency response ± 3dB Po = 2 watts 50Hz-25KHz. Sensitivity for Rated O(P—Vs = 25v. RL = 80 ohm. f=1KHz 75mV. RMS. Size: 75mm x 63mm x 25mm,

AL20; 5W £2.95 AL30 10W £3.25

Specially designed for use in-Disco Units, P.A. Systems, high power Hi-Fi, Sound reinforcement systems

Y £15.95+8%

POWER AMP

1-5A

P-O-W-E-R

AND for those who need more

SPECIFICATION:

AL250

Output Power: 125 watt RMS Continuous

Operating voltage: 50-80

Loads: 4-16 ohms

Frequency response: 25Hz-20kHz Measured at 100 watts Sensitivity for 100 watts output

at 1kHz: 450mV Input impedance: 33K ohms Total harmonic distortion 50 watts into 4 ohms: 0.1% 50 watts into 8 ohms: 0.06% S/N ratio: better than 80dBs Damping factor, 8 ohms: 65 Semiconductor complement: 13 transistors 5 diodes

125

Overall size: Heatsink width 190mm, length 205mm, height 40mm

£6.70

NEW PA12 Stereo Pre-Amplifier completely redesigned for use with AL25-36 Amplifier Modules. Fea-tures include on/off volume. Balance, Bass and Trebie controls. Complete with tape output. Frequency Response 20Hz-20KHz (-3dB) Bass and Trebie range±12dB Input Impedance 1 meg ohm Input Sensitivity 300mV Supply requirements 24V. SmA Size 152mm x 84mm x 33mm



supply for AL20-30. PA12. S450 etc. Input voltage 15-20v A.C. Output voltage 22-30v D.C. Output Current 800 mA Max, Size 60mm x 43mm x 26mm.



COMPONENT SHOP: 18 BALDOCK STREET, WARE

You lucky people!

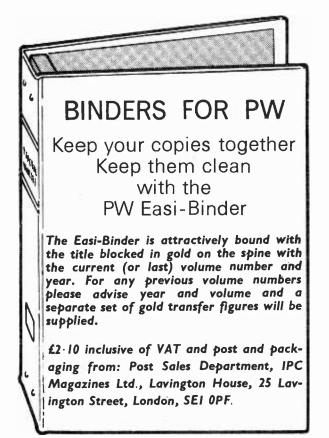
Y ES, dear reader, if you have just paid your 40p for this copy of PW then you have really got your money's worth this month! Apart from the free information card contained in this issue you will also find a free copy of the index for Volume 52 of PW which finished with the April issue. 'Fine' you may say, 'So what'? Well, it is something new for PW because in the past the index was produced some time after the end of the volume had been reached and if you wanted to get a copy of it you had to pay 45p! Plus all the bother of finding writing paper, envelope, stamp and postal order or cheque!

One would need to go back into the past quite a bit to discover how this practise arose in the first place, and more to the point, why it has persisted. Anyway, that's all history now and, hopefully, you'll get your free index on time in future.

Another relic of the past also needs to be dealt with and that is to make our volume of twelve issues coincide with the calendar year! I have no intention of wading through early volumes of PW to discover just when this discrepancy occurred but in the early days of PW in 1932 it was a weekly publication and the subsequent change to a monthly appearance probably explains the mystery.

I have often wondered whether those conscientious readers of PW who bind their copies do so by the volume or by the year. Personally, mine go into binders by the year but the office copies are very neatly bound by the volume! Ah, well, perhaps we will get it all together 'ere long. I don't suppose these little matters bother our average reader in the slightest, but it was something to talk about!

Eric Dowdeswell Assistant Editor





More outlets for trio

U P to the present time, Trio High Fidelity equipment has only been obtainable from selected dealers within the UK. A move has now been successfully made to enable all Trio equipment to be available from Comet Discount Warehouses. Of course the same equipment will still be available from the selected dealers, but as from 1st February of this year, they have been redesignated as "Trio Specialists'. B. H. Morris and Co (Radio) Ltd., Precision Centre, Heather Park Drive, Wembley, Middx HA0 ISU.

Bib book

LANNED, and written for by Clement beginners Brown, the BIB Book Of Hi-Fi explains in laymens terms the basics of Hi-Fi, stereo principles and the source of high quality programmes, and offers hints and tips to those planning new systems. Installation and adjustments are explained in a practical form, and emphasis is placed on good modern practice in equipment maintenance and record and cassette care. A glossary of Hi-Fi terms and an index conclude this latest offering from BIB. Priced at £1.98, it can be obtained from Hi-Fi dealers, bookshops, and audio components specialists.

Bib Hi-Fi Accessories Ltd, Kelsey House, Wood Lane End, Hemel Hempstead, Herts.

The author of the Gas/Smoke Sensor Alarm—April '77, acknowledges the help of WKF Electronics Ltd., Welbeck St., Whitwell, Worksop, Notts., in the construction of the Alarm. WKF can supply all components for this project.

VACANCY ON PW

Applications are invited for the post of TECHNICAL EDITOR on P.W. which will become vacant shortly.

Apply, with brief details of career, to Eric Dowdeswell, Assistant Editor.

HiFi '7'7

A LREADY 74 exhibitors have booked space for the fourth annual HIGH FIDELITY exhibition, which is to be held at the Heathrow Hotel, London Airport, from the 19th April to the 24th April, 1977.

NEW

Special events taking place during the exhibition will include a series of lectures and demonstrations in the Hotel's York Video Theatre, while facilities are being claimed to be 'second to none'. These will include airconditioned and soundproofed exhibition rooms; free car parking; exhibition catalogue and free between the free transport Heathrow Hotel and London Airport terminals and Hatton Cross Underground Station.

Opening times for High Fidelity '77 will be as follows:

- Trade and Press only—19th to 21st April inclusive 10am to 8pm.
- Public---22nd to 24th April inclusive 11am to 9pm (with the exception of the 24th which closes 6pm.)

Admission for everyone is free. Further information from Emberworth Ltd., London House, Oxford Road, Stokenchurch, Bucks. Tel: 024026 2674/5.

Bits and pieces

RADIO clubs in the UK may be interested to learn that one of our readers has acquired a large amount of prewar and war-time radio equipment and wishes to dispose of most of it. Any clubs that would be interested please contact M. Robinson, 13 Tolmers Gardens, Cuffley, Potters Bar, Herts.

PA Exhibition

T F PA Equipment is of interest to you, the Association of Public Address Engineers will be holding their annual exhibition at the New Wembley Conference Centre, from April 19th to the 21st. The displays will be located on the ground floor and will contain some of the most sophisticated and up-to-date equipment available in the World. Amongst them will be alarm systems, amplifiers, background music systems, sport event timing equipment and general PA and audio equipment.

NEWS

The exhibition is open each day from 10am to 6pm (5pm on last day). Admission is free and "anyone having a professional or business interest is welcome to attend", say the organisers.

Further information from APAE HQ, 47 Windsor Road, Slough, Berks. Tel: Slough 39455.

Roll up, roll up . . .

N EW members are required for the Chester and District Amateur Radio Society, whose newly formed committee is formulating a new

Aerial guides

FOR many years now the general public and electronic enthusiasts have realised the importance of a good aerial system, to utilise the full potential of modern tuners and radio receiving gear.

In the forefront of modern aerial development is Antiference, who have recently published a Wall Chart and FM Stereo guide. The Wall Chart measures $23^{1}_{2in} \times$ 16^{1}_{2in} and is printed in bright yellow, red and black and gives a complete up-to-date guide to all UHF TV transmitters, together with all the main Antiference TV aerial products.



programme for the coming months. Meetings are held at the YMCA Chester on the 2nd, 3rd and 4th Tuesdays of each month at 8pm.

Any readers interested in joining the club should contact G. W. Chaliner GW8DMR, 34 Chestnut Avenue, Summerhill, Wrexham, Clwyd.

Rally call

THE North Midlands Mobile Rally will take place on Sunday 17th April at Drayton Manor Park, near Tamworth, Staffordshire. This event is organised by the Midland and Stoke-on-Trent Amateur Radio Societies, and will include the usual exhibitions.

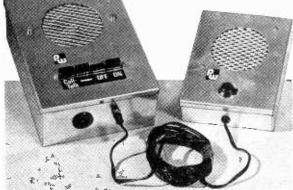
Further information, stickers, badges etc., from Barrie Willetts G8DEM, Alf Walton G3ZKQ or Max Gutteridge G8BHE at 68 Max Road, Quinton, Birmingham.

The FM Stereo Guide comprises a three colour pocket folder illustrating the complete range of "Mushkiller" VHF/FM aerials with individual polar diagrams and comparative gain graph. The FM Station guide gives a complete "run down" on BBC and IBA transmitters showing clearly the polarisation, ERP in kW and frequency of transmissions. A map of the UK is also incorporated to show the relative positions of the transmitters, together with detailed photographs of aerial accessories.

Further information from Antiference Ltd., Aylesbury, Bucks. Tel: 0296 82511.







THE author required a two-way intercom for his workshop, which would also permit listening for the telephone, doorbell, burglars, etc, when no-one else was at home. Suitable amplifier designs which have been published all seem to have a high input impedance. Since it was intended to use the loudspeakers as microphones, this would have necessitated a transformer to match the loudspeakers to the amplifier input. The circuit described here eliminates this need.

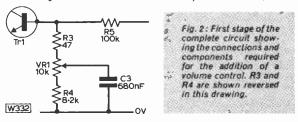
The Circuit

The amplifier circuit is shown in Fig. 1. Tr2-5 form a conventional push-pull amplifier, with the quiescent current in the output transistors being set by D1 and D2. The input stage Tr1 however, is connected in the common base configuration, which gives the required low input impedance, with high voltage

Steven DAVIES

gain. Negative feedback is applied from the output to the base of Tr1.

The DC voltage at the emitters of Tr4 and 5 is set by R3, although for optimum performance this voltage should be half supply voltage, which could be obtained by substituting a $10k\Omega$ preset and $12k\Omega$ resistor in series for R3. This is not strictly necessary here, since HiFi is not really needed. The AC gain is set by the ratio of R5 and R4, and if a volume

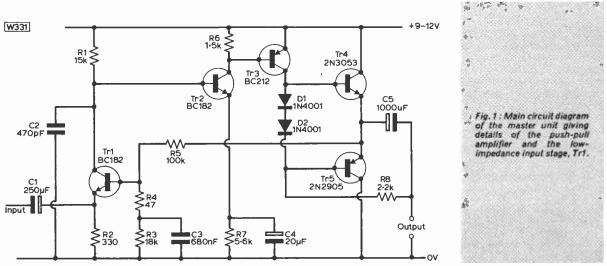


control is required, the circuit of Fig. 2 can be used, which simply varies the amount of AC feedback applied.

Capacitor C3 has been chosen to roll off amplifier response at low frequencies to reduce pickup of mains frequency. C2 rolls of response at high frequencies to eliminate RF pickup, which can be a problem when long lines are used. (Radio 2 on 1500m is the main offender.)

Operation

To communicate, S2 should be in the "ON" position, Fig. 3, while the direction of communication is



Practical Wireless, May 1977

★ components list

			and the second second			
Resis	fors					
R1	15kΩ		R6	1·5kΩ		
R2			R7	5·6kΩ		
R3			R8	2.2kΩ		
R4			R9	680Ω		
R5			VR1	10kΩ		
	W 5%		• • • •			
A						
Capa	tors					
C1	250µF	15V	C5	1000µF	15V	
C2			C6	100µF	15V	
C3			C7	2000#F	15V	
C4	20µF	15V	0.	2000,0		
64	LOAI	134				
Sami	conduct	075				
Tr1	BC182	013	D2	1N4001		
Tr2			D3	1N4001		
Tr3			D4	1N4001		
Tr4			D5	1N4001		
Tr5			D6	1N4001		
D1	1N4001		00	1144001		
U	1144001					

Miscellaneous

S1, CPDT biased switch. S2, SPST switch. S3, SPST switch. S4, DPST switch. LS1, 8–35 Ω loud-speaker. LS2, 8–35 Ω loudspeaker. T1, mains transformer with 7 to 8V 200mA secondary. Clip-on heat sinks for Tr4 and 5. Aluminium or wooden cases to suit. PCB, 120 x 60mm (available from the PCB readers Service page 25). F1, 2A fuse. NE1, mains neon. Wire, plugs.

determined by the talk/listen switch S1, one speaker acting as a microphone. To conserve power, S2 will normally be "OFF" so that there is then no DC path between supply 0V and amplifier 0V, due to the capacitor C6 at the slave unit. If a DC path is made, either by shorting out C6 with S3, or by operating S1, the circuit will oscillate, the resulting tone being heard in both speakers.

The cause of oscillation is not at first sight clear. Consider Fig. 4, which is a block diagram of the system when "calling". Initially C5 will be discharged. When power is supplied to the circuit, C5 will start to charge up, via Tr4, LS1 and LS2. The

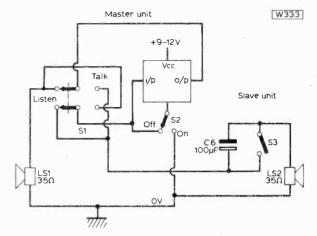


Fig. 3: Schematic, demonstrating the wiring and switching arrangements required between the master and slave units.

resulting voltage across LS2 will increase the input voltage, and therefore the voltage drop across LS2, will fall to almost zero. This will cause a negativegoing signal at the amplifier input, switching Tr5 on and discharging C5 via LS1. The cycle then starts again, producing oscillations.

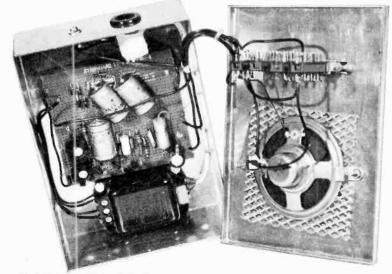
Power Supply

The circuit can be operated off a 9V battery, but for prolonged or frequent use a mains supply is recommended. Any supply of 9-12V DC at 200mA minimum will suffice. The prototype used a small 8V bell transformer T1, with rectification and smoothing as shown in Fig. 5. Note that if a mains supply is used any exposed metalwork should be earthed, and the live line fused at a maximum of 2A, either in the plug or in the unit itself.

Construction and Installation

No problems should be experienced with construction, as all components are housed on a single PCB as shown in Fig. 6.

The cases were standard aluminium boxes, $101 \times 132 \times 38$ mm for the slave, and $177 \times 127 \times 63$ for the



General interior photograph of the master unit. As this is the prototype, construction was on Veroboard and not a PCB as described in the text.

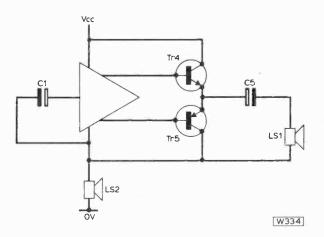


Fig. 4: Block diagram of the system in the 'calling' mode, showing how the tone is produced.

master. Wooden boxes are a possible alternative, since screening is not necessary. Using smaller speakers than specified would permit a smaller box,

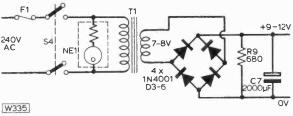


Fig. 5: Alternative power supply for the intercom. The transformer in the prototype was an 8V bell transformer, although any transformer rated at 9-12V 200mA would suffice.

Switch S1 should be biased, and may be difficult to obtain, although such switches are often available on the surplus market. Alternatively, it is possible to "doctor" a bank of push buttons to provide S1 and S2, as in the prototype (note that one of the banks does nothing, so three would suffice).

Once made, the units should be mounted on a convenient wall, and joined with two-core cable (screened cable could be used, but is unnecessary). Take care to get the polarity right, otherwise C3

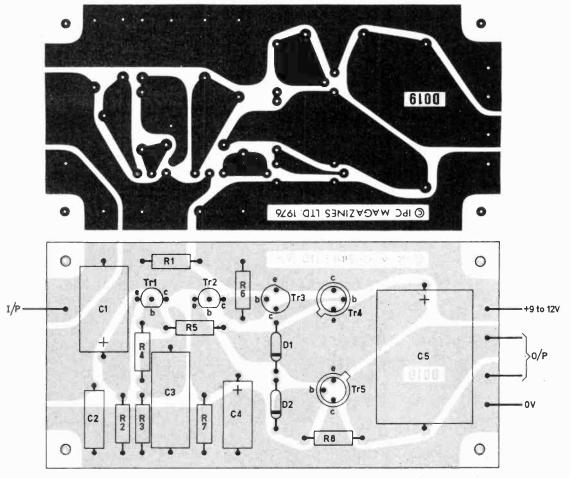


Fig. 6: Drawing of both the foil and component side of the PCB designed for this project.

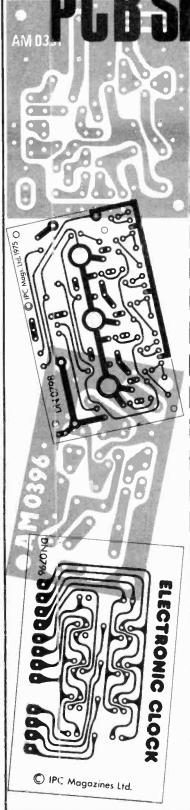
but unless space is critical there is little point in this. The amplifier will drive speakers down to 8Ω , so there should be no problem in obtaining something suitable.

might suffer. It is advisable to route the cable away from mains as far as possible, to prevent undue pickup, since the hum level in the amplifier itself is very low. All boards, except SRBP, are glassfibre, drilled and roller-tinne

To:- READERS PCB SERVICES LTD, PO BOX 11, WORKSOP, NOTTS

Please supply PCB/s as indicated by tick/s in box/es

	Issue	Project		Ref	Price P/	P
	Sep 75 Dec 75 Dec 75	Electronic Clock (set of three) Sound-To-Light Display Disco System, Amplifier		DN0795/6/7 DN0798	$2 \cdot 40 + 15$ $1 \cdot 15 + 12$	
	Dec 75 Jan 76	(2 required) Disco System, Light Modulator Music Box	each SRBP	A M0421 A M0423 D N1/JM	$4 \cdot 40 + 22$ $3 \cdot 50 + 22$ $2 \cdot 25 + 18$	
		CMOS Crystal Calibrator Wobbulator Auto. Slide Synchroniser Dig. Freq. Meter (set of 5)	lassfibre A015 an	DN1/JM AM0438 AM0443 AM0441 d 4x A004	$3 \cdot 00 + 18$ $1 \cdot 19 + 12$ $1 \cdot 08 + 12$ $2 \cdot 33 + 15$ $3 \cdot 17 + 15$	
	July 76	Transistor Tester Disco Preamplifier Cassette Player Power Supply Capacitance Meter Digital Car Clock (set)		A002 A003 A001 A009 A001/2/3	$3 \cdot 08 + 18$ $0 \cdot 65 + 12$ $0 \cdot 65 + 12$ $2 \cdot 59 + 14$ $2 \cdot 58 + 12$	
	Oct 76 Oct 76 Oct 76 Nov 76 Nov 76 Nov 76	Interwipe Video-Writer (set) Hazard Flasher Low Level Battery Indicator Electronic Thermostat Cirtest Probe	D002/3	DN8JM 8/4/6 A007 D005 A016 A017 A018	0.80 + 12 21.44 + 50 0.76 + 12 0.40 + 12 1.30 + 12 0.48 + 12	
	Nov 76 Dec 76 Jan 77 Jan 77 Jan 77	Burglar Alarm Chromachase Oscilloscope Calibrator Icelert Polyphon, motor and main boar		A019 A021 A023 A020 A025/4 A008*	0.50+12 5.70+22 1.25+12 1.45+12 7.90+20 0.90+15	
	Feb 77 Mar 77 Apr 77 Apr 77 May 77	Polyphon, tune disc, blank, SRE Transistor Checker FM Stereo Touch Tuner Tug 'o' War (set) Gas/Smoke Sensor Alarm 2-Way Intercom	54	A026 D023/4/5 A029/030 A028 D019	$1 \cdot 18 + 12$ $7 \cdot 50 + 20$ $2 \cdot 88 + 12$ $0 \cdot 65 + 12$ $1 \cdot 28 + 12$	
71	May 77 May 77	Protected Battery Charger Seekit Metal Locator	cot of b	A027 A031	2·38+12 3·38+12	
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Part 1:-FIXED REGULATORS

A LARGE number of integrated circuit voltage regulator devices are now available which enable very stable, high performance power supplies to be constructed very easily with the simplest possible circuitry. Such regulators are extremely valuable when one requires a very stable output voltage as the load current or as the input supply voltage varies. They are equally useful when the hum level from a mains power supply unit must be kept to a very low level. For example, a few mV of hum on the tuning voltage feeding a varicap front end will produce hum at the FM detector output, but a small monolithic regulator in the varicap supply can be used to prevent this.

This article has been written to guide readers on the general principles of operation of regulator devices and to provide information on the types readily available for the home constructor and the basic circuits in which they can be used.

DEVICE CLASSIFICATION

Monolithic regulator devices can be divided into two main classes, namely fixed and variable output types. The former is basically designed to give a certain specified output voltage, although it is possible to increase this voltage somewhat whilst keeping a very simple circuit, whilst the latter gives an output voltage which can be varied by means of a potentiometer in the external circuit. Regulators can also be classified according to whether they are designed to provide a positive or a negative output voltage. However, these types are to some extent interchangeable.

Classification can be according to the maximum rated output current. The maximum current from the low current devices is about 100 to 200mA; such devices are normally mounted in a transistor type package and seldom require a heat sink. Medium current devices provide maximum outputs of 200mA to 1.5A and will usually be used with a heat sink, whilst devices delivering still higher currents are always used with a substantial heat sink.

Low current regulators can be used in more complex circuits with external power transistors to provide quite high output currents. This approach is often more economical than the use of high current monolithic or hybrid devices which are normally far more expensive than the low or medium current devices.

PACKAGES

The low current regulators are normally supplied in transistor or dual-in-line packages. Many medium current devices are in plastic packages with a tab or metal insert for mounting the device on a heat sink, whilst high current devices are often encapsulated in the diamond shaped TO-3 metal package used for power transistors. The recent trend is for the use of low current "on card" devices, the regulator being on the same PCB as the devices to which it supplies power and this greatly reduces unwanted coupling due to the line impedance.

The general principles of operation of a voltage regulator are shown in Fig. 1 which uses a 741 op-amp as a comparator. The zener diode D1 provides a reference voltage V_{ref} and the 741 compares this with the voltage tapped off by VR1. If the value of V_{ref} is the greater of the two, the 741 output will be high and Tr1 will conduct so that the output voltage, V_0 , rises to a value which brings the potential of the slider of VR1 to V_{ref} the output of the 741 falls and Tr1 presents a higher impedance. Thus the output voltage is stabilised at a value higher than that of V_{ref} the value being determined by the setting of VR1.

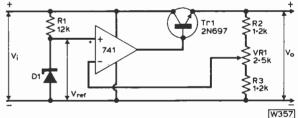


Fig. 1 : A voltage regulator using a 741 operational amplifier.

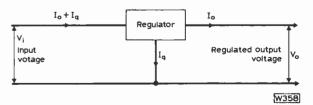


Fig. 2: Regulator currents and voltages. I_{σ} is the quiescent current which flows even when the output current is zero.

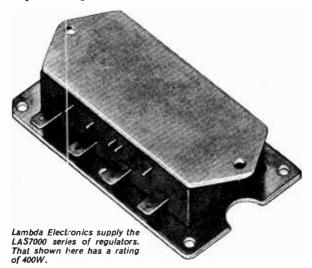
The 741 input requires little current and therefore does not affect the stability of the V_{ref} voltage. Most voltage regulators operate on this principle, with all the components fabricated on a single chip and usually including protective circuits. The maximum output current is determined partly by the ratings of the series transistor Tr1.

POWER DISSIPATION

The power dissipated internally in a regulator may be estimated using the circuit of Fig. 2. A quiescent current I_i flows through the device at all times to enable it to operate correctly, usually of the order of 10mA. If no output current is taken, the internal power dissipation is $V_i I_q$ and is quite small owing to the low value of I_q . When an output current is taken there is an additional internal dissipation of $(V_i - V_o)I_o$ since $(V_1 - V_o)$ is the potential difference through which I_o falls as it flows through the device. Thus the total dissipation is $(V_1 - V_o)I_o + V_i I_q$.

INPUT VOLTAGE

What input voltage should one apply to a regulator? The maximum value is quoted on the data sheet and damage may occur if it is exceeded even for a millisecond! The minimum input voltage should always be a few volts above the output voltage, since this additional small voltage is required to enable the device to operate correctly. The minimum value of $(V_1 - V_0)$ is known as the 'drop out' voltage, since at lower voltages the regulator 'drops out' and ceases to provide regulation.



As the internal power dissipation is $(V_i - V_o)I_o + V_iI_q$, it is very desirable to keep V_i as low as possible in order to minimise the dissipation. This is especially important in the case of high current regulators where I_o , and therefore the internal dissipation, is large. In the case of variable regulator circuits which provide a wide range of output voltages, V_i must exceed the maximum value of V_o so the value of $(V_i - V_o)$ and the dissipation can be large when the output voltage is set to a low value. If the output current is moderately high, it is advisable to consider the use of a switching regulator circuit which can offer higher power efficiency at the expense of circuit complexity.

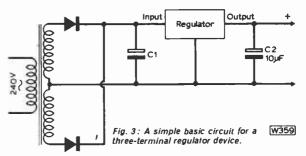
PROTECTION CIRCUITS

Most modern regulator devices incorporate circuits to protect them from damage if the output is shorted or if the device becomes too hot, the former by a circuit which limits the maximum value of the output current to a safe value. Thermal protection can take two possible forms. In one type an increase of the chip temperature will cause the output current to be reduced so that the temperature cannot increase further and cause damage to the device. Nevertheless, it is bad practice to operate a regulator for a considerable time at such a high temperature that this current limiting circuit operates; such high temperatures tend to reduce device reliability.

Another form of protective circuit used in low to medium current devices is that of foldback current limiting. If a low resistance is connected across the output or if the output is shorted, the current 'folds back' to a much smaller value than the maximum output. Thus the internal power dissipation is kept relatively small.

BASIC CIRCUIT

The basic circuit used with simple three terminal voltage regulator devices is shown in Fig. 3. The input voltage is developed across the reservoir capacitor C1 and the regulator provides a constant output voltage. In addition, the hum level at the



output is far less than that across C1.

If a slightly increased output voltage is required, a resistor may be included between the 'earth' terminal and the zero voltage line. For example, if this resistor has a value of 270Ω and the quiescent current is 10mA, the output voltage will be raised by the voltage across this resistor, namely $2 \cdot 7V$. Alternatively a zener diode may be use instead of a resistor.

LOW CURRENT DEVICES

One of the best known series of fixed output low current regulators is the TBA625A (5V), the TBA435 ($8 \cdot 5V$), the TBA625B (12V) and the TBA625C (15V) series produced by SGS-ATES Ltd. (available from Chromasonic Electronics). They can be used in the very simple circuit of Fig. 3 to provide output currents of up to about 120mA. The devices incorporate foldback current limiting; if the output is shorted, the current falls to between 30 and 45mA. In addition, the short circuit current falls with increasing chip temperature, to provide further protection.

These regulators reduce the hum level across C1 by a factor of about 400 times (52dB). The larger the value of C1 the smaller the hum voltage across it and the lower the output hum level. If C1 is about 250μ F the output hum level will be of the order of 1mV. C2 may be required with some devices to prevent instability and should be close to the device. The dropout voltage is about $2 \cdot 3V$ at the maximum current, whilst the output impedance is of the order of $0 \cdot 1\Omega$.

This series of regulator devices is available in the

type of transistor package shown in Fig. 4(a). The devices may be used in the type of circuit shown in Fig. 5 to provide a small range of output voltage variation. In this circuit the output voltage is equal to $V(1+R2/R1)+I_qR_2$ where V is the voltage of the regulator itself. The circuit of Fig. 3 can also be used to provide a negative supply if the positive line is earthed instead of the negative line.

THERMAL LIMITING

Another type of 100mA regulator is made by Fairchild as the μ A78L00 series, by Motorola as the MC78L00C series and by National Semiconductor as the LM78LXX series, but the latter has been replaced by the improved LM340L-XX devices. When referring to a specific device in the series, the '00' or 'XX' is replaced by the voltage of that device. For example, the μ A78L05, the MC78L05C and the LM3400L-05 are 5V devices, whilst the LM340L-24 is a 24V device.

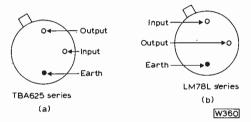


Fig. 4: Connections for (a) the TBA625 series and (b) the LM78LXX series.

These devices are available in TO-5 metal cans, but the connections, Fig. 4(b) are different from those of the TBA625 series; they are also available in TO-92 plastic packages. Some types are stocked by Chromasonic Electronics Ltd., whilst National Semiconductor devices are available from DTV Group Ltd., 126 Hamilton Rd., London S.E.27.

These devices can be used in the same circuits as those already discussed. The output current on short circuit does not fold back, but is limited in value and a thermal limiting circuit operates at high chip temperatures. The performance is similar to that of the TBA625 series, but the dropout voltage (1.7V)is rather less.

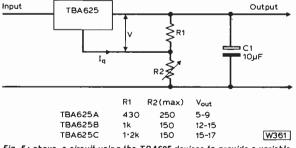
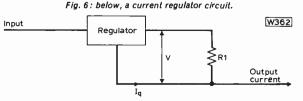


Fig. 5: above, a circuit using the TBA625 devices to provide a variable regulated output.



Voltage regulators of almost any type can be used as a current regulator, in the circuit of Fig. 6. The output current is held constant at a value of $V/R1+I_q$ where V is the output voltage of the device concerned.

MEDIUM CURRENT DEVICES

The National Semiconductor LM342-XX series has similar internal circuitry to the LM340L-XX series. but can supply in excess of 200mA. The SGS-ATES TDA1405 (5V), TDA1412 (12V) and TDA1415(15V) are available from Doram Electronics Ltd, and have a similar internal circuit to the TBA625 series, but can deliver output currents of up to 500mA. They are encapsulated in a 3-lead plastic package with a hole for bolting the metal insert to a heat sink. The foldback limiting causes the current to fall to between 80 and 180mA on short circuiting the output. However, thermal damage can be caused if the device is not mounted on a heat sink. The L129 (5V), L130 (12V) and L131 (15V) are similar to the TDA1405 devices, but have a wider operating temperature range. All these devices can be used in the type of circuit already discussed.

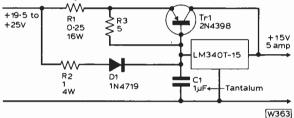


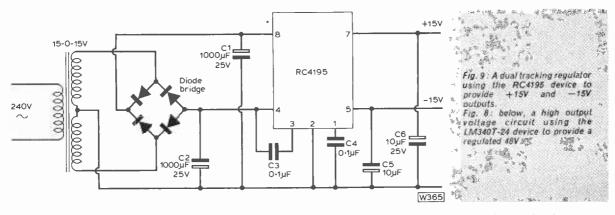
Fig. 7: A circuit for increasing the output current from a LM340T-15 device by a factor of five.

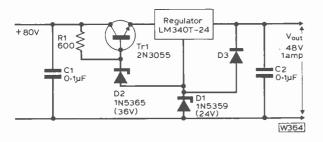
Another series of medium current devices has been adopted by many manufacturers, thermal limiting being employed rather than foldback limiting. The internal circuitry is similar to that of the LM342-XX series but the output current can be up to 500mA in some types and up to 1A in others. These standard types are available with nominal outputs of 5V, 6V, 8V, 12V, 15V, 18V and 24V. For example, the 1A range includes the LM340T-XX in a TO-220 plastic package, the LM340K-XX in a TO-3 package, the Motorola MC7800 series, the Fairchild #A7800 devices and the Texas Instruments SN72900 series. The 500mA devices include the National Semiconductor LM341P-XX plastic types with a tab for bolting to a heat sink and the Fairchild µA78M00C. Other devices available from Doram Electronics Ltd include the LM309K 5V regulator in a TO-3 package which can give over 1A and the LM323K in the same package for 5V at up to 3A.

HIGHER CURRENTS

All these devices can be used in the circuits discussed, but the type of circuit shown in Fig. 7 can be used to obtain more output current. In the particular circuit shown, a LM340T-15 device is used, the maximum current being 5A instead of the 1A of the device itself.

If the voltage across D1 is equal to the base-emitter voltage of Tr1, the current flowing through R1 is about four times that through R2. This current boosting circuit takes advantage of the internal current





limiting properties of the regulator to provide shortcircuit protection of the booster section. The short circuit is limited to R2/R1 times the output limiting current of the regulator device itself. If the heat sink fitted to Tr1 has at least four times the capacity of the sink fitted to the LM340T-15 device, the thermal overload protection of the latter will be extended to Tr1. As the load current increases from 0 to 5A, there is a change of about 1% in the output voltage.

HIGH OUTPUT VOLTAGE

The circuit of Fig. 8 shows how the LM340T-24 can be used to provide an output of 48V. Under shortcircuit conditions the voltage at the input side of the regulator falls to 35V, D3 aids the start-up of the circuit and protects the regulator from high inputto-output voltage differentials during short circuits.

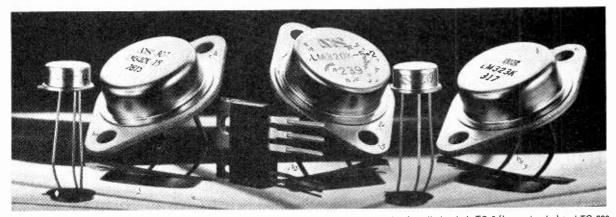
DUAL TRACKING REGULATORS

Work with operational amplifiers normally requires dual power supplies of about $\pm 15V$ and the supplies should be regulated to prevent the output voltage of the operational amplifiers from drifting. Further, it is desirable for accurate work that the drift in the voltage of one of the pair of lines should be matched by a proportional drift in the voltage of the other line; that is, the two voltages must track one another.

One of the simplest ways of constructing a dual tracking power supply unit with $\pm 15V$ outputs involves the use of the Raytheon RC4195 device in the circuit of Fig. 9. This device is available from Doram Electronics Ltd (order code 303-636) as an 8-pin dual-in-line package which can deliver over 100mA from each output with short circuit currents of about 220mA. The input voltages should be in the range $\pm 18V$ up to the maximum of $\pm 30V$. The frequency compensating capacitors C3 and C4 increase stability, but can often be omitted.

The RC4195 device incorporates thermal shut down and short circuit protection for both outputs. It replaces two separate regulator devices. The balance connection to pin 6 can be taken to a variable voltage so that the two output voltages can be accurately balanced. The RC4195 can also be used to provide a single regulated output of $\pm 50V$ at 100mA, whilst in another circuit with external power transistors it will provide a dual tracking supply of $\pm 15V$ at 2.5A each.

Next month, in Part 2, we shall deal with Fixed Regulators



This photograph, provided by National Semiconductor, shows voltage regulator devices in TO-5 (small circular), TO-3 (large circular) and TO-220 plastic packages.



Radio what?

Next time you are in the company of people who delight in using long words, get your own back; say radioimmunoassay! If they call your bluff and ask what it means, the following explanation might help.

The word radioimmunoassay (RIA for short) comes from the medical profession, well known for its love of technical tongue twisters. RIA is a technique used to measure very very tiny amounts of things in the blood such as hormones etc. The medical minnions often do this by putting in some minute but radio active tracer elements and following their path around the body. These tracers are often measured while still in the body.

A rather bright medical instrument manufacturer has now gone one better. The antigens (these are the minute traces you want to measure) are bound to microscopic flecks of iron. It is then, in crude lay terms, simply a matter of going fishing with a magnet! The fluid is passed by a magnet and the iron particles are separated leaving only the elements in the sample which are required for measurement. In practical terms this means that the manual separation part of RIA can be fully automated.

Micro speaker

Beepers for use as alarms have been marketed for some time and are available in this country from various electronics stockists. However, although their idea is not new I am very taken with the very latest one just launched in America.

The beeper measures only 0.64in cube and weighs just 6 grams. Despite its very small size, the device has a free field output of 105dB at 1in. The secret of this great output compared to size is in the cunning construction of the inside of the cube which houses the device. This internal structure is fashioned like a miniature folded horn which is tuned to 2.1kHz. An advantage of a tuned horn as opposed to a resonant cavity is that the horn tunes very broadly and so is not critical. Yet another advantage is that the beeper, because of its broader bandwidth, can be used as a truly miniature loudspeaker. It is offered in a variety

of impedances from 2Ω to $2k\Omega$ and it even has little solder tabs for ease of mounting onto a PCB. One can imagine a very small radio set for, say, the bedroom using a ZN414, LM380 and one of these tiny speakers. Alas—they've only just been released onto the market in America and are not obtainable in this country as yet.

Expensive games

With television games gaining greater popularity one would think that this was a real growth area for electronics and without too many problems. Sad to tell this is not the case. I hear of a report which indicates that the grid or pattern projected onto the CRT of a television (i.e. the goals for football or the walls for squash etc) is "burning" a permanent image onto the TV tube and, as such, will eventually ruin it for normal TV watching. The time taken for this effect to become noticeable is, so I'm told, about 2.000 hours of use. However, with different games, different intensity settings and, indeed, different tubes, this time is by no means a standard to judge the situation by.

Now, a group has been set up to study this affect and to report its findings to the authorities. I will mention this again as and when more news comes in. However, the thought of ruining a colour television tube is a little frightening when one thinks of the cost. Perhaps, if you play TV games, it might be prudent to keep the brightness turned down just a little?

Straight tape

If you drill a hole how do you know if it's straight? Probably this is unimportant for your ordinary average hole, but if your drilling real big deep ones, like oil people do, then it becomes very important. A British company has manufactured what is called an inertial navigation unit (more long words). In use, this unit is lowered down the hole. On its journey it keeps careful records; it has a cassette recorder to monitor and log the distance it has travelled and deviation in terms of lateral movements. When the unit is hoisted

ON RECENT DEVELOPMENTS

up the tape is played to an analyser unit which in turn feeds out printed data on the accuracy of the hole at different depths.

Before you decide to use one for the precision drilling of this years potato crop, the price for one such installation is quoted as around £91,000—and that's a 'hole' lot of money.

Hot news

If you wanted to take the temperature of something it's almost certain that you'd use some kind of thermometer. I see that the market has just been invaded by another non-contact thermometer and it struck me that some readers may not have heard of these intrigueing devices.

This latest one is a small hand-held unit which is simply pointed at the target. It will then, within one second, show the true temperature of that target in degrees C. The target can be up to 20 feet away and this particular "thermometer" has a measurement range from 10°C right up to 1,000°C.

It occurs to me that if I was only 20 feet from 1,000°C I would become medium rare in about 20 seconds!

More expensive?

When hire transport in London changed from handsome cab to motor taxi there must have been some regretful sighs. I wonder if there will be any sighs for taxi meters. A problem is that a cab may be laid up for a matter of days while its electromechanical meter is adjusted. Owners of a few of the larger fleets of cabs have got together and formed a company to manufacturer an electronic taxi meter. Readout of the fare cost is, of course, digital while changing the meter is just a matter of unplugging one and plugging in another.

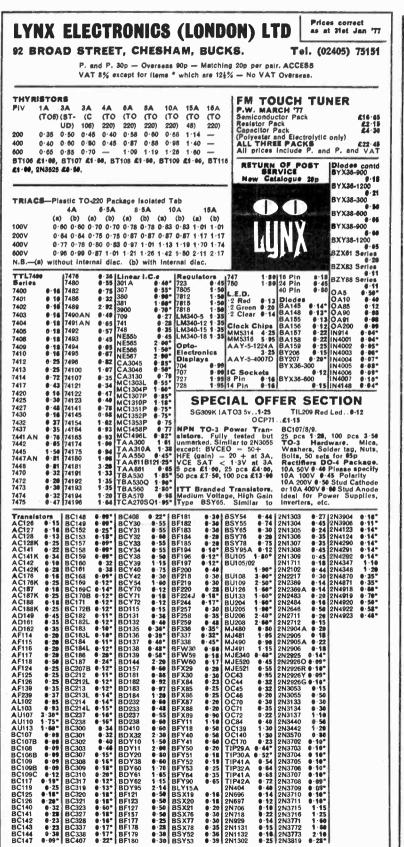
Watch out for the new best seller folks, "Electronic Taxi Meters—Are They Really Fare?"



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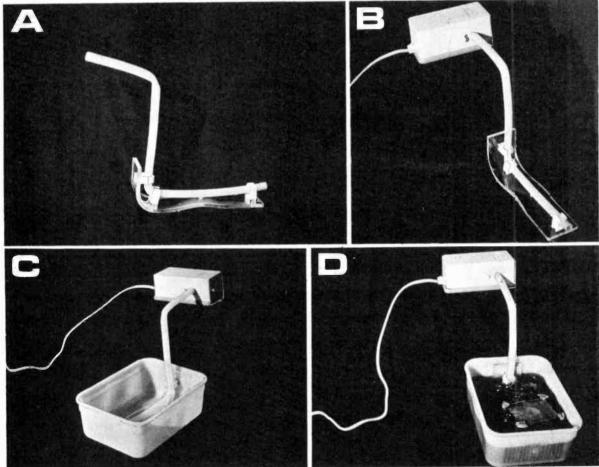
improved JA.KENNEDY etching of PCB's

ANY chemicals and combinations of chemicals are suitable for etching copper but by far the most common, due to its cheapness and ease of use, is ferric chloride, either by itself or mixed with various additives. In solution ferric chloride reacts with the water to form a small amount of the insoluble salt ferric hydroxide, together with hydrochloric acid (HCl). This reaction is easily prevented by adding up to 5% HCl. In either case HCl is present and available for reaction with both ferrous chloride and cuprous chloride, converting them back to ferric chloride and cupric chloride respectively if free oxygen is present. FeCl₃ + $3H_2O \rightarrow Fe(OH)_3 + 3HCl$ Hydrolysis. FeCl₃ + $Cu \rightarrow FeCl_2 + CuCl$ At copper surface. FeCl₃ + $CuCl \rightarrow FeCl_2 + CuCl_2$ In body of solution. CuCl₂ + $Cu \rightarrow 2CuCl$ At high copper levels. If oxygen is present:

 $\begin{array}{l} 4\mathrm{FeCl}_2 + \mathrm{O}_2 + 4\mathrm{HCl} \longrightarrow 4\mathrm{FeCl}_3 + 2\mathrm{H}_2\mathrm{O} \\ 2\mathrm{CuCl} + \mathrm{O}_2 + 2\mathrm{HCl} \longrightarrow 2\mathrm{CuCl}_2 + 2\mathrm{H}_2\mathrm{O} \end{array}$

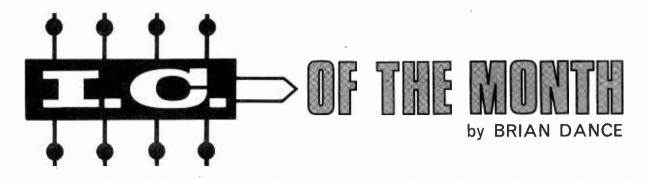
From the above it can be seen that it is advantageous to have an oxygenated solution. The two commonest methods of oxygenating a solution (spraying it through air or bubbling air through it) also provide mechanical agitation at the copper surface thus removing the ferrous and cuprous chlorides as they are formed and supplying fresh etchant to the surface.

Spray etching produces better results than bubble etching but the problems involved in producing a uniform spray pattern and obtaining pumps manufactured in materials that are inert to the etchant tend to cancel any advantages to the amateur. In comparison, bubble etching is very simple to imple-



Photograph A shows the plastic tube suitably bent and fixed to a heavier plastic base. In B the tube has been fitted to the air pump. Photograph C shows the complete set-up with the tube in the bottom of a plastic

bowl. Finally in \mathbf{D} the solution is seen bubbling with the PCB floating on the surface with the aid of a cork in each corner.

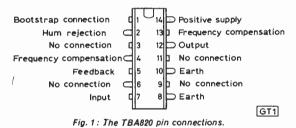


Number 62

SGS-ATES TBA820 LV AUDIO AMPLIFIER

THE SGS-ATES TBA820 device is not designed to give a very high output power but is very suitable for use in portable equipment where the battery voltage may be as low as 3V. The typical quiescent current consumption is only 4mA at 9V, so one can obtain a long battery life at low output power.

The TBA820 is very suitable for use in small portable radio receivers, small record players and tape recorders, etc. It enables much simpler audio power amplifiers to be constructed compared to those where only discrete components are used.



The TBA820 is encapsulated in a quad-in-line package with the connections shown in Fig. 1. In this type of device alternate pins on each side of the plastic body are bent so that they protrude to different distances at their lower ends. This configuration provides more space between adjacent pins for ease of soldering than in dual-in-line devices, but quad-in-line sockets are more difficult to obtain.

CIRCUITS

A typical circuit for the TBA820 is shown in Fig. 2. If the input signal has any DC superimposed on it, a capacitor of value about 0.22μ F must be placed in series with the input connection. R1 may be replaced with a $100 k\Omega$ volume control, if desired, but a resistor of value not exceeding a few hundred $k\Omega$ must be present from pin 7 to earth to achieve correct biasing.

The capacitor C6 prevents a steady current flowing from the output (pin 12) through the loudspeaker speech coil. A somewhat smaller value can be used, but this will result in a reduction in the bass response. The power supply decoupling capacitor, C4, should be soldered fairly close to pin 14 to prevent spurious RF oscillation. C7 is an electrolytic capacitor in parallel with C4 which provides decoupling at lower frequencies.

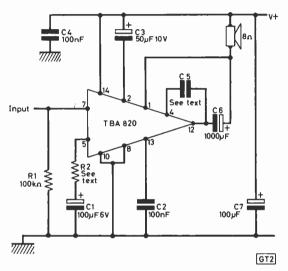


Fig. 2 : A simple TBA820 amplifier circuit.

The gain of the circuit is determined by the value of R2. An internal feedback resistor of value $6k\Omega$ is connected from pin 12 to pin 5, so the smaller the value of R2, the greater the gain. The gain is approximately equal to 6000/R2; thus voltage gains of 30, 60 and 120 may be obtained by using values of R2 of 200Ω , 100Ω and 50Ω respectively. The maximum value of R2 is about 390Ω if a 9V supply is employed or 270Ω with a 12V supply, since saturation of the input circuit may occur with higher values.

The value of C5 determines the high frequency response of the circuit, but the value required for any specified frequency response is affected by the value of R2. The value of C5 required for various upper cut-off frequencies is plotted against R2 in Fig. 3.

The optional capacitor C3 provides a reduction of about 42dB of any hum on the power supply line. It is not required if a well stabilised supply is employed. However, if a battery is used as a source of power, this capacitor should always be included, since it will prevent 'motor boating' when the internal resistance of the battery rises with age.

The maximum output power which can be obtained from TBA820 depends on the impedance of the loudspeaker and on the power supply voltage. For example, one can obtain 2W with an 8Ω speaker

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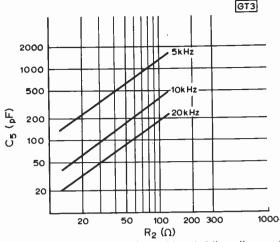
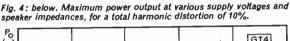
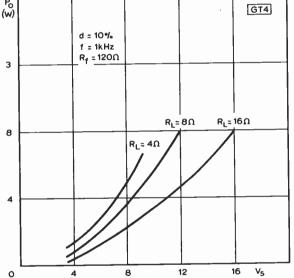


Fig. 3: above. This graph may be used to select the optimum value of C5.





and a 12V supply. The absolute maximum power supply voltage is 16V which will enable 2W to be obtained from a 16 Ω speaker. However, it is wise to regard 14V as the maximum to allow a reasonable margin of safety. If a 4 Ω speaker is employed, the power supply voltage should not exceed 9V to prevent overloading of the device. An output of up to 1.6W can then be obtained. If the circuit is operated from a 9V supply when using an 8 Ω speaker, the maximum output is about 1.2W, Fig. 4.

If the supply voltage is only 3.5V, the maximum output power is about 0.2W. Although this may seem very small, the ear has a logarithmic response and this power level is quite adequate for a portable receiver; indeed, many small speakers cannot handle more. The writer found that the circuit shown will operate satisfactorily with power supply voltages down to 3V but at still lower voltages the distortion increases rapidly and the gain falls. At low supply voltages a relatively low impedance loudspeaker of 4Ω may be chosen so as to provide a reasonable output power.

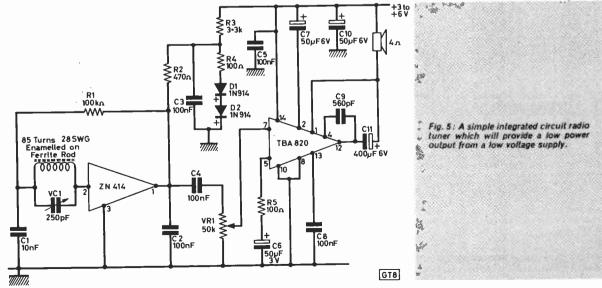
Cross-over distortion is not present in the TBA820. As the power output increases, the distortion remains almost constant at a low value of about 0.2per cent until clipping of the waveform begins. The distortion then rises very rapidly with increasing output power.

The TBA820 can dissipate up to 1.25W. In practice this means that no heat sink is required when the device is used with power supplies of up to 16V with a 16 Ω loudspeaker, up to 12V with an 8 Ω loudspeaker or up to 9V with a 4 Ω loudspeaker.

RADIO RECEIVERS

The writer has used a TBA651 AM radio circuit (see *Practical Wireless July 1974*) which fed a signal into a TBA820 power amplifier. This circuit used an 11V supply but a different integrated circuit must be employed in the radio section in receivers using a low power supply voltage.

A simple low voltage radio is shown in Fig. 5; only medium wave coverage is provided by the circuit shown, but a long wave coil could be switched into



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the circuit if desired. About 220 turns of 40 SWG insulated wire is suitable for a long wave coil. The two silicon diodes D1 and D2 keep the supply voltage to the ZN414 fairly constant as the power supply voltage falls with an ageing battery. Low voltage capacitors are used in the circuit of Fig. 6, since they are smaller and cheaper than higher voltage types. However, the power supply voltage must not exceed 6V when these capacitors are used.

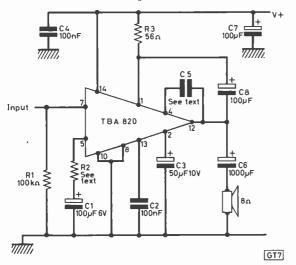
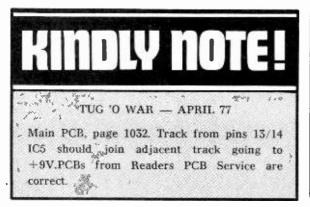


Fig. 6: A circuit in which one side of the speaker can be earthed.

IN USE

The TBA820 may be soldered in position on a piece of matrix board with 0.1in hole spacing. However, a neater amplifier can be made by using a printed circuit board. It should be noted that the input earth connection should be separate from the output earth connection, to minimise unwanted coupling.

If the constructor wishes to employ a circuit in which the loudspeaker is earthed on one side, the circuit shown in Fig. 6 may be used. This does require an additional electrolytic capacitor C8 to provide the bootstrap signal to pin 1. The component values not marked should be selected in the same way as the corresponding components of Fig. 2. If a hum rejection or "anti-motorboating" capacitor is employed in this circuit, it should be connected between pin 2 and earth and not as in Fig. 2.



IMPROVED ETCHING OF PCB's-contd from page 33

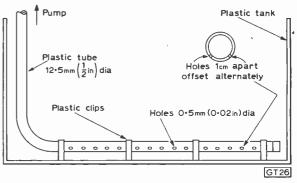


Illustration of the plastic tube showing how the holes in the tube are drilled.

ment, as it increases the useful life of the etchant and can reduce etching time to less than five minutes thus minimising any tendency for the resist to deteriorate. All that is required in addition to the etching tank is a high capacity air pump as used for aquariums and a length of ${}^{1}2''$ dia. PVC tubing. The pump should be capable of maintaining a pressure of 2lb/sq.in.

CONSTRUCTION

The tube is mounted close to the bottom of the tank as shown in the diagram with 0.5mm holes spaced about 1cm apart on the underside of the tube offset alternately by about 45°. The end of the tube in the tank is sealed with a rubber bung and the other end attached to the pump, the pump being mounted above the level of the etchant.

The etch rate is greatest near the surface of the etchant therefore the best results are obtained by floating the board copper side down on the surface of the etchant. This is most easily achieved by cutting the board slightly oversize and inserting the corners into small corks. Double sided boards should be etched vertically. If it is desired to increase the etch rate still further an aquarium heater set to about 30°C can be used to heat the etchant.

A suitable etching solution would be: Ferric Chloride (FeCl₃) 500 gm. 35% Hydrochloric Acid (HCl) 50ml. Water (H₂O) 1 litre.

NOTES

Up to 1ml of photographic wetting agent may be added but this may cause foaming which could prove troublesome. When in use adequate ventilation must be provided and the solution should not be left in an open tank any longer than necessary as it will promote oxidisation of any metal in the room.

The most suitable method of disposing of the ferric chloride is to neutralise it with calcium carbonate (limestone) and then bury it. (N.B. hard water should not be used for making the solution.)

The success or failure of PCB production depends to a great extent on the cleanliness of the copper being maintained at all times. An economical medium for mechanically cleaning the board is a 'Scotchbrite' pan scouring pad followed by a wipe with a tissue wetted with acetone in order to remove any grease present. Acetone is also suitable for removing 'Dalo' type resists.



DAVID GIBSON



'The Ohm Gnome'

MANY of the previous circuits in this series have found major applications in novelty. The project described here has a more serious role.

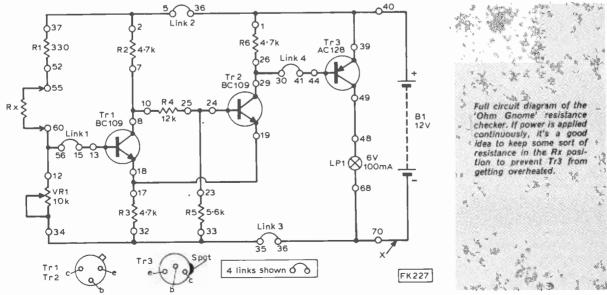
Basically it is a device for indicating the value of "unknown" resistors. To this end it employs rather novel principles—and no expensive meter. In use, the resistor whose value is unknown (or in doubt) is plugged into S-DeC holes 55 and 60. The potentiometer, VR1, is rotated until the bulb lights. The pot is then rotated slowly in the opposite direction until the lamp goes out, and the value of resistance is then read or interpolated from the calibrated dial.

Circuit ideas

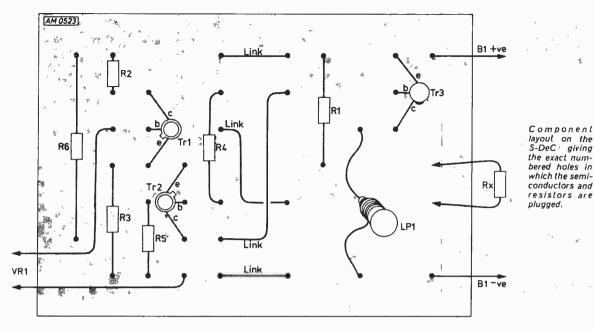
The most immediate thought in designing such a circuit would be to use a simple bridge of some sort,

you will need

1 x x x x	*	1 - Y	lu the the
R1 330Ω	Tr1	BC109	8.30
R2 4-7kΩ	Tr2	BC109	
A R3 4 7kQ	Tr3	AC128	
R4 12kΩ	LP1	6V 100mA bulb	
R5 5-6kΩ	B1	12V battery	4
-R6 4·7kΩ	One S-	DeC	× *
- VR1 - 10kΩ pot -	One Blo	b Board ZB-5D	
walk / w	*	And States	



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and indeed circuitry exists which advocates this. However, despite the most careful tests this idea was rejected. To turn the potentiometer until the lamp was extinguished was almost impossible without using a magnifying glass to be really sure that the bulb was out or at least at "minimum red glow".

In the circuit diagram, transistor Tr3 drives LP1 upon receipt of a suitable signal from Tr2 and so the Tr3 part of the circuit is quite straight forward. Let's look at the role played by Tr1 and Tr2.

These transistors are connected as a Schmitt trigger circuit. This useful arrangement can be used for our purpose because it triggers at a reasonably precise level/value of input signal.

Potential divider

In the circuit we are varying the input by rotating the potentiometer. The latter, together with R1 and the unknown value of resistor under measurement forms a potential divider. The setting of VR1 at which the circuit triggers will vary depending upon the value of the unknown resistor. Thus by calibrating the potentiometer dial using known values, the unknown ones can be read directly or interpolated.

The circuit will work well by simply "tuning" for the bulb to switch on, but it seemed easier on test to set the pot more precisely by "tuning" for bulboff. The only real difference in practical terms is that the dial readings are slightly displaced between the two sets of readings. Providing, when calibrating, the constructor settles for one method or the other and sticks to it (i.e. the bulb-on or-off giving indication) then either method is entirely satisfactory. Using a $10k\Omega$ potentiometer for VR1 gave a measurement range from 33Ω to just over $33k\Omega$. The useful measurement range was found to extend from $1k\Omega$ to $33k\Omega$ because the lower end of the range (below $1k\Omega$) tended to be very cramped and difficult to read with any degree of accuracy. Substituting a $1M\Omega$ potentiometer for VR1 gave a measurement range from about 33 Ω to 2.2M Ω but the dial was rather cramped.

Just plug-in

Assembling the circuit on the S-DeC is extremely simple and easy. Just plug in the components via their own leads. Push the leads into the S-DeC hole number(s) given in the circuit and check again, against the full size S-DeC layout drawing. Using the little jumper leads with plugs on each end is helpful when links are required.

For those constructors who would like to make a permanent unit of the Ohm Gnome the following suggestions are offered. The simplest method of construction, once the circuit has been "proved" on the S-DeC, is to transfer the components directly onto a piece of BloB Board. Because all the components are soldered and fitted on one side of the board only, the BloB Board may be stuck (with any good adhesive) directly to the inside of a suitable case.

A push button should be inserted between the negative terminal of the battery and the negative line of the circuit (point X in the circuit). In operation, press the push button and hold it down. Rotate the pot until LP1 lights, then release the push button (if tuning for lamp-off, hold push button until you've "tuned" back to extinguish the light). In this way the battery life will be considerably extended since the circuit will only draw current when actually measuring.

Dial construction

With a reasonable size case, a useful size of dial can be employed, and it is suggested that the constructor consider making the dial with the aid of a protractor. Mark the dial around the potentiometer pointer accurately, in degrees. Then, using a piece of graph paper mark degrees along the X axis, and the known resistance values along the Y axis. Plug in, say, a $1k\Omega$ resistor, $2 \cdot 2k\Omega$, $4 \cdot 7k\Omega$ $10k\Omega$ $12k\Omega$ $22k\Omega$ $27k\Omega$ $33k\Omega$. These will give you all the points you need to draw a graph. You will find most of these resistors from previous projects in this series.



CW FILTER UNIT VERSATILE AF GENERATOR

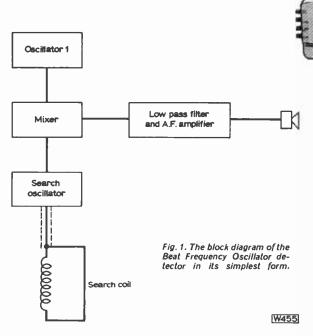
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A

5

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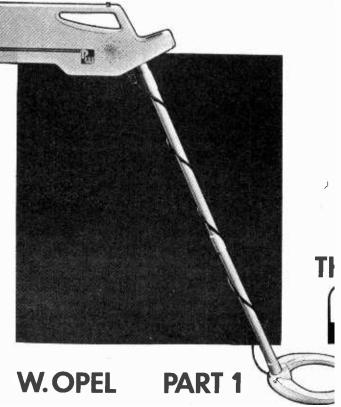


RUE Beat Frequency Oscillator detectors operate on the principle that the outputs of two separate oscillators are combined to pro-.duce an audio tone although the oscillators may be at non-audible frequencies.

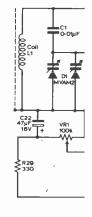
The metal locator uses the search coil as the tuned circuit of one oscillator so the presence of metal objects modify the resonant circuit causing a change in the beat note. Fig. 1 shows the block diagram of a typical detector.

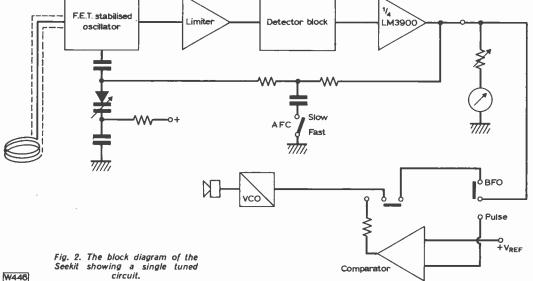
Although this type of detector is easy to construct it does have drawbacks, namely:---

- 1. Metering facilities are ineffective and insensitive. 2. The two oscillators may drift in different direc-
- tions and make operation tedious. 3. It is difficult to obtain high detection sensitivity
- combined with stable (drift free) operation.
- 4. In general the battery life is proportional to the output volume and, since the output is usually a sine wave, high outputs are required.









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circuit.

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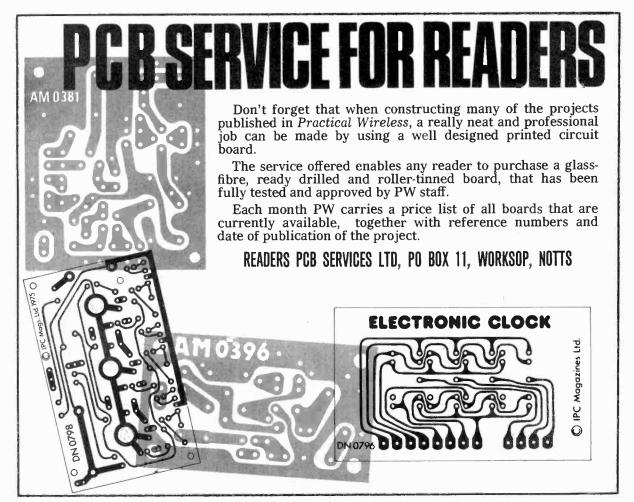
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CIRCUIT OPERATION

This new unit overcomes the first two of these problems by using a single search oscillator whose output is fed into a discriminator coil and the frequency changes are converted to voltage variations in a similar manner to an FM tuner. Fig. 2 is the block diagram of this unit and Fig. 3 is the full circuit diagram.

Since only one oscillator is employed it becomes that much more effective to apply AFC to the oscillator to cure drift and improve the long term tuning stability. To accomplish this effectively, the small voltage differential at the detector output is amplified in a Norton Amplifier (an LM3900) and fed to the oscillator, via a variable time constant (R17 and C15), and applied to D2. Provision is made to switch C15 in or out in order to give two sensitivity settings.



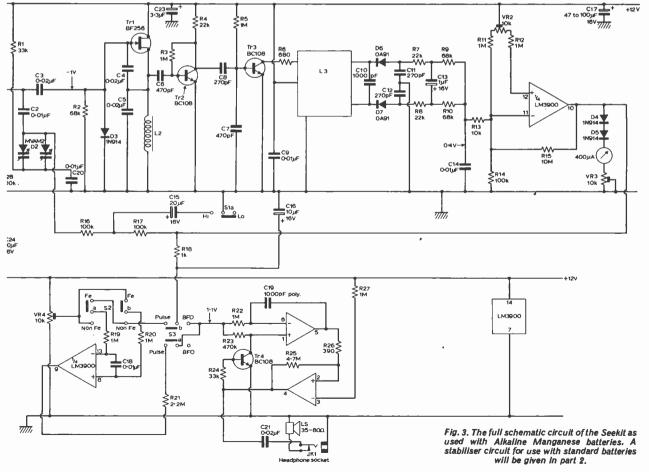
With C15 in circuit the high output at pin 10 of IC1, resulting from the presence of a metalic object entering the field of the search coil, takes an appreciable time to effect the oscillator and the detector is at maximum sensitivity to varying frequency fluctuations.

With C15 switched out the output of ICIa is fed direct to the oscillator and input variations produce virtually instantaneous AFC correction. In this mode the range is effectively halved.

The output of ICla is sufficient to drive a sensitive meter. This meter provides an earlier warning of the presence of metal objects since the eye can detect smaller changes than the ear.

ICla also drives the audio amplifier circuit, based on IClc and d, when the unit is switched to BFO. The output is available either through a speaker or plug-in earphones. This audio is in the form of a square wave and produces a higher audibility than a sine wave consuming the same power thus prolonging battery life.

The main tuning of the unit is performed by the varicap diode D1 with resolution afforded by VR1, a 22 turn preset potentiometer. Because the oscillator utilises a very high L/C ratio it is important to use a heavy gauge wire for the search coil (the prototype used 22 SWG) to maintain the "Q". This also means that the large amount of capacity used to tune the circuit results in relatively small frequency variations, even with the abrupt type of varactor used.



With the unit switched to "Pulse" it is possible to adjust VR4 such that IClb is just below the threshold of conduction. Under this condition the output of ICIa is fed to one input of ICIb but, because of the setting on the other input, the audio amplifier is off. When a signal is present at the search coil it results in a change at pin 10 of IC1 and ICb switches to the on state. Its output is fed to the input of the audio stages and the treasure is advertised. This mode has the advantage that the detector can be used on the speaker output without annoying others present.

In conjunction with the pulse operation is switching for Ferrous or Non-Ferrous detection. The presence of ferrous (iron based) objects will lower the frequency of the tuned circuit whilst nonferrous will increase it. In the Seekit type of detection this will result in deviations in opposite directions, see Fig. 4, and by switching from ferrous to non-ferrous, whilst in the pulse mode, it is possible to decide if the object is what we are looking for. Unfortunately, at the frequencies used, large pieces of ferrous material can cause the detector to react as if it had located something non-ferrous due to HF eddycurrents set up within the material. Although a sweep of the area will indicate whether the find is large or small it could still be a large non-ferrous item.

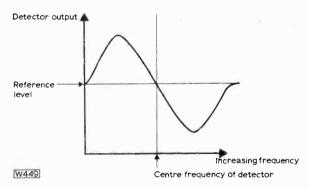


Fig. 4. The changes in output voltage resulting from frequency changes. Use is made of the positive and negative changes to identify ferrous or non-ferrous materials.

THE SEARCH HEAD

The capability of the unit is largely a function of the search coil. Small coils are best for detecting small objects fairly near the surface whilst large diameter coils will sweep a large area and detect

Coil	Number	Inductance	Frequency
diameter	of turns		with ·01µF
			Capacitor
7″ (175 mm)	25	243μH	102kHz
	24	224 µH	106kHz
8″ (200 mm)	23	235µH	104kHz
	22	215µH	108kHz
9″ (225 mm)	22	242μH	102kHz
	21	220μH	107kHz
10″ (250 mm)	21	245µH	101 kHz
	20	222 µ H	107kHz
1			

Table 1. The approximate resonant frequencies for coils of varying diameter and number of turns.

★ components list

Resistors	C4	0.02µF poly-
R1 33kΩ		carbonate/Mylar
R2 68kΩ	C5	0.02µF poly-
R3 1MΩ		carbonate/Mylar
R4 22kΩ R5 1MΩ	C6	470pF ceramic/
R5 1MΩ		polystyrene
R6 680Ω	C7	470pF ceramic/
R7 22kΩ		polystyrene
R8 22kΩ	C8	270pF ceramic/
R9 68kΩ		polystyrene
R10 68kΩ	C9	0.01 µF ceramic/
R11 1MΩ	1.1	polystyrene
R12 1MΩ	C10	1000pF polystyrene
R13 10kΩ	C11	
R14 100kΩ		270pF ceramic
R15 10MΩ		1µF 16∨
R16 100kΩ	C14	
R17 100kΩ	C15	
R18 1kΩ		10µF 16∨
R19 1MΩ	C17	47 to 100µF 16V
R20 1MΩ	C18	01µF ceramic
R21 2·2MΩ	C19	
R22 1MΩ	C20	
R23 470kΩ		·02µF
R24 33kΩ		47µF 16∨
R25 4.7MΩ	C23	3.3µF 16V
R26 390kΩ	C24	
R27 1MΩ		polycarbonate where
R28 100kΩ R29 330Ω	spec	cified
All 1W 5%		
VR1 100kΩ multiturn		
type AB47	Sen	niconductors
VR2 10kΩ preset type	IC1	LM3900
PT10V	Tr1	8F256 (FET)
VR3 10kΩ preset type	Tr2	BC108 or similar
PT10V	Tr3	BC108 or similar
VR4 10kΩ preset type	Tr4	BC108 or similar
PT10V	D1	MVAM2, MVAM115
	D2	MVAM2, MVAM115
Capacitors	D3	1N914, 1N4148
C1 0.01. E coramic	D4	1N914, 1N4148
C2 0.01μ F ceramic	D5	1N914, 1N4148
C3 0.02µF poly-	D6	0A91, 0A90
carbonate/Mylar	D7	0A91, 0A90
Inductors		
L1 search coil, see text		
L2 20 to 43 mH screened	choke	e, Toko type 10RA/RB
L3 Detector assembly, h	nigh Q	, Toko type ST3/D2
Section Contract Sectors		
Miscellaneous		
Speaker, 35 to 80Ω. Cryst	tal ear	piece with plug and
socket. 4 way change-over	switc	h, push button, non-
interlocking. Meter, edge	readir	ng, 400µA or better.
Case, moulded or suitable	e box	to hold components.
Detector head, moulded or		
and joints if required. Co		ig and socket. Coax
cable, approx 1 metre (3ft)	. –	

objects at a greater depth. Thus the coil used will depend to a large extent on the type of hunt.

Whatever size of coil is used it is essential to exercise care in its construction, details of a satisfactory method are given later. The varying numbers of turns for various diameter coils are given as Table 1. These coils have been calculated to give an operating frequency of about 90kHz since this is within the approved range of the Home Office and has advantages over lower frequencies in the band.

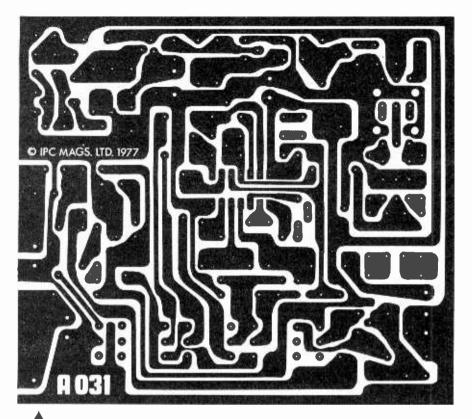
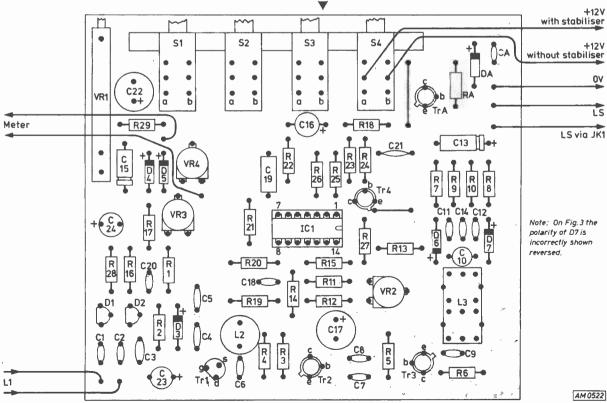


Fig. 5. The wiring pattern of the electronics board, shown full size to enable direct copying for the 'Do-It-Yourselfer'.

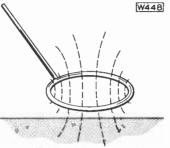
Fig. 6. The component positions and orientions. Note that provision is made for the stabiliser required if standard batteries are to be used over long periods.



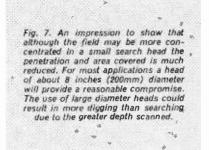
If searching for a particular object recently lost (e.g. coins on the beach or a ring in the garden) a small coil is required but if combing an old battle field looking for swords etc., then the large coil is the answer, see Fig. 7. The 8 inch (200mm) coil on the prototypes was able to detect a 2 pence piece at between 8 and 10 inches in free air and, with a well shielded coil, in most other non-metalic materials.

to the individual constructor but the following points must be observed.

- 1. The wire should be 20 or 22 SWG.
- 2. The windings must be wound tight and held tight after winding.
- 3. A Faraday electrostatic shield should be fitted but remember that the shield must not be a complete short circuit around the coil.



Larger coil offers better depth of penetration, though less ability to pinpoint smaller objects



Smaller coil concentrates field and assists pinpointing

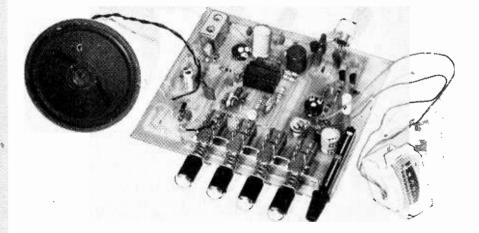
CONSTRUCTION

Apart from the speaker, the phone socket, the meter, the batteries and the search coil all the components are mounted on a single printed circuit board. Fig. 5 shows the wiring pattern of the board whilst Fig. 6 shows the component location and orientation on the board.

The prototypes were constructed as follows and took care of the above conditions.

Firstly several strips of adhesive tape were fastened to a saucepan of about 7 inches diameter such that a portion of the sticky side faced outwards. Next, 25 turns of 20 SWG enamelled copper wire, with a quality lacquer coating rather than a polyurethane finish, were wound tightly around the former

Fig. 8. A photograph of the prototype board, complete with the stabiliser circuit components. Comparison with the component overlay of Fig. 6 will show some changes in the components used. These were necessary to improve long-term stability of the oscillator. The second prototype, constructed to Fig. 6, was working when photographs were needed.



The meter and an output socket are mounted on the case of the unit and, if the moulded case is used, the mounting positions are cast in during manufacture.

The batteries are fixed into standard holders and the holders themselves are bolted to the case.

The electronic items do not present any problems in their assembly since it is largely a matter of wiring-up a printed circuit board. The coil, on the other hand, does require very careful construction. The details of a suitable former etc., can be left and on the tape. When the winding was completed the ends were held tightly whilst the tape was folded over the turns. The winding was slid off the former and the tapes bound tightly around the turns, covering as much of the winding as possible. Extra tape should be wound over the wire to cover it completely. The tape should be wound as tight as possible and if heat shrink tape is available it could be used with advantage.

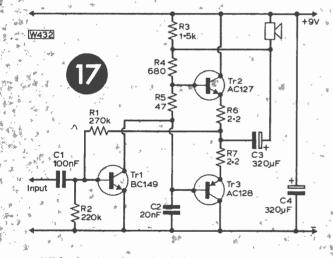
This article will be completed next month.

Circuits for AUDIO AMPLIFIERS

PART 3

COMPLEMENTARY SYMMETRICAL OUTPUT

Fig. 17 is a more convenient circuit, using a single battery. The speaker is coupled by C3. Each output transistor has its own emitter resistor, to help to stabilise working conditions. Overall stabilisation is obtained by R1 from the output emitter circuit back to Trl base. DC feedback of a similar type is found in many driver/output circuits of this kind. It is necessary to use the transistors shown (or near equivalents) or the value of R5, in particular, may need changing.



With circuits of this kind, best results are obtained with the correct speaker impedance. Somewhat higher impedances may be used with a loss of volume. Lower impedances should not be used, as this may damage the output stage transistors.

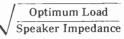
SPEAKER IMPEDANCE

The speaker impedance specified for a circuit should be used, depending as it does on operating conditions and power output expected. A lower impedance may cause peak currents to be exceeded or result in too high a dissipation in output devices. A higher impedance than specified usually causes reduced power output and possibly other troubles. Changes to output load can upset feedback circuits and the correct load is particularly necessary with high fidelity equipment.

Practical Wireless, May 1977

F.G.RAYER G3OGR

Where a matching transformer is used, its ratio is obtained from the formula:—



For example, a 3 Ω speaker is operated from an output stage requiring a 255 Ω load, so $\sqrt{255/3}=9\cdot2:1$ approx. These are the operating conditions for the amplifier in Fig.14.

The transformer must be able to handle the required power, from a few hundred milliwatts with small amplifiers up to hundreds of watts with large amplifiers. Windings must be of low DC resistance for the rated power to be realised.

THERMISTOR COMPENSATION

It was seen that acceptable working conditions are easily maintained over a range of temperatures and voltages in the case of low power amplifiers where emitter stabilisation was possible. In other circumstances additional means of stabilisation are needed, and Fig.12 included a thermistor in a Class A stage.

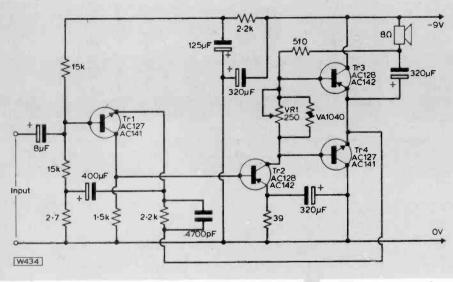
Fig.18 is a 1W transformerless amplifier in which the VA1040 thermistor stabilises the operating current of the output pair Tr3 and Tr4. The resistance of the VA1040 falls as temperature rises, thus reducing the base/emitter bias voltage of the pair. This effect is the opposite of that arising in the transistors themselves.

Quiescent (no signal) current is set by VR1 to between 12 and 15mA. The emitter of Tr1 samples the DC voltage at the emitters of Tr3/Tr4, and by DC coupling through Tr2, maintains DC operating conditions. Tr3 and Tr4 require heat sinks about 4×4 cm of 16SWG aluminium, or larger. DC operating conditions of all stages are related, so values should be as shown.

QUIESCENT CURRENT

The "silent" or quiescent current of an output stage is very easily adjusted with circuits such as Fig.19, as DC operating conditions are isolated from other stages. R1 and R2 set the base bias point and their relative values control the collector current of Tr1 and Tr2. Reducing the value of R1, or increasing the value of R2, moves the base bias negative and increases collector current (with PNP devices). Reducing R2 or increasing R1 has the opposite effect.

If the quiescent current is low, objectionable crossover distortion arises, especially with falling supply voltage or reduced temperature. If current is high,



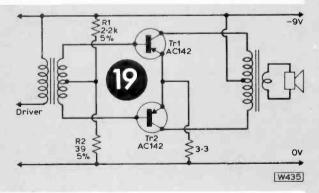


A 1W four-transistor amplifter in which a thermistor VA1040 is used to stabilise the output pair.

dissipation in the transistors rises, so close-tolerance resistors are often necessary for R1 and R2. An alternative is to use a pre-set resistor, such as 100Ω for R2. The total collector current can then be set to 8 to 10mA, for the transistors shown.

INTEGRATED CIRCUITS

Numerous ICs suitable for audio applications are available and they can be used in a wide range of circuits. They may be obtained with gain, powerhandling and other ratings suitable for pre-amplifier, output and other purposes. Some require relatively few extra discrete components (resistors and capacitors) and they are available for a considerable range of operating voltages. Some have integral overload protection. These devices may be used alone or in conjunction with transistors and can often offer advantages compared with the use of individual transistors for all stages.

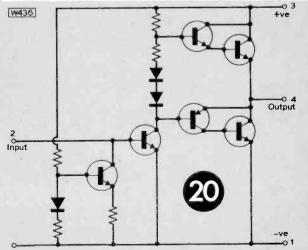


Amplifiers using ICs may be assembled on perforated board or PCBs prepared for the purpose. Some ICs have tabs which must be soldered to areas of the board, for heat sinking. This in turn, may depend on the operating voltage, or power required. For example, the TBA800 audio IC requires no heat sink provision, other than the tabs provided on it, for up to 1W but for higher outputs the tabs are soldered to copper areas of the PC board. Other ICs rely on the use of certain pins, or alternative methods, of carrying heat away from the interior of the device. With many ICs there are optional speaker loads, depending on the operating voltage and power dissipation. The TBA800 is also an example of this and can be expected to deliver up to about 1.8W with a 160 speaker but about 3W with an 80 speaker, when using 14V. Apart from the obvious advantage of using a pair of ICs for stereo, ICs may be combined in arrangements similar to the push-pull output stage, allowing more power than would be available from a single IC of the same type.

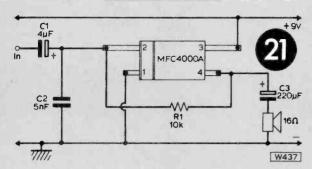
THE MFC4000A

This is a simple, low-power audio IC having the configuration shown in Fig.20, employing stabilised driver and output stages. There are four external connections only, as shown. Fig. 21 is an operating circuit for this IC the extra components consisting of an input coupling capacitor C1, by-pass capacitor C2, speaker coupling capacitor C3, and feedback resistor R1.

The more important ratings for an audio IC are similar to those which would apply to an amplifier constructed from discrete components. These characteristics allow an IC to be chosen to suit the



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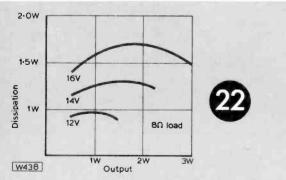
purpose in view. For the MFC4000A typical ratings are:

Operating voltage	9V
No signal current	3·5mA
Maximum output	350mW
Sensitivity (50mW output)	15 mV
Harmonic distortion (50mW)	0.7%

If complete details of an IC are required, or an internal circuit (such as that in Fig.20) these are often available from the maker or supplier. Otherwise it is more general to show only the external circuit (Fig.21) as this is all that is required to build an amplifier. In Fig.21, the top of the device is shown, with pins in their actual locations. With other circuits, pins are numbered, and do not usually appear in their actual locations. The latter are found by reference to a diagram of the device.

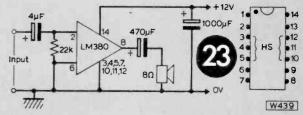
VOLTAGE/POWER

With an IC amplifier there is a range of possible operating voltages. The lower limit is set by the device ceasing to give an adequate performance (or to operate at all) while the upper limit depends on the device itself, heat sink and speaker impedance. With many circuits less than maximum operating voltage is used, but the voltage should only be increased when it has been checked that operating conditions are still suitable.

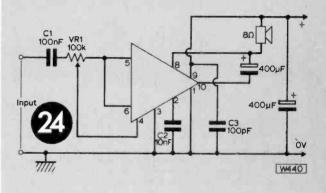


In Fig.22 the dissipation expected in the LM380 IC is shown for 12, 14 and 16V, with an 8Ω speaker. When no heat sink is fitted, dissipation should be kept under 1.25W. This sets 14V as a maximum safe voltage with an output of a little over 2W. With 12V about 1.5W output would be expected. With 16V a heat sink is necessary.

Fig.23 is a circuit for the LM380 and it also shows the pins. When a heat sink is necessary, pins marked HS are soldered to it and it may be a foil area of the PCB. If high frequency oscillation arises, a 1000μ F capacitor with a 2.7Ω resistor in series may be



connected from pin 8 to negative line. The 1000μ F capacitor should have short leads to pin 14 and the negative line. Top-cut tone controls may be connected from pin 2 to 6.



SL414A/SL415A

These IC's provide 3W or 5W, with 18V or 24V using an 8Ω speaker, the power dropping to approximately 2.2W and 3.8W with a 16Ω speaker. The IC includes a pre-amplifier (24dB) and main amplifier (26dB) sections and automatic overload protection. Fig.24 is a simple circuit for these ICs with input for the main amplifier (4) taken from the volume control VR1. Satisfactory working, with reduced output, is possible down to 9V.

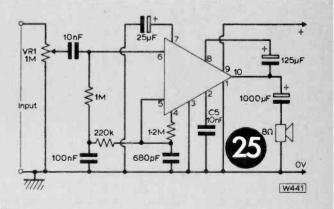


Fig.25 is also suitable for the SL414A or SL415A. More components are required but the pre-amplifier output (5) is taken to the main amplifier input (4) and gain is increased. A high-frequency by-pass capacitor may be required from pin 9 to pin 1 near the IC (as in Fig.24). In some cases a 22 Ω resistor and 47nF capacitor in series may also be required from pin 10 to pin 1.

TO BE CONTINUED





THIS battery charger offers several features which are usually omitted on the general run of charger circuits.

In particular, it protects itself and the battery from damage in the event of the charging leads being connected the wrong way round, it protects itself against damage from the charging leads shorting together, its charging rate and the end voltage of the battery are easily adjusted during manufacture or subsequently and the components are relatively easy to obtain.

DESIGN IDEA

To provide the above features the author chose a circuit using a constant current generator and a variable on/off period of charging. The switching for the on/off is a mixture of electronics and electromechanics since the charging current is switched by a thyristor whilst the thyristor itself is switched by a pair of contacts on a relay.

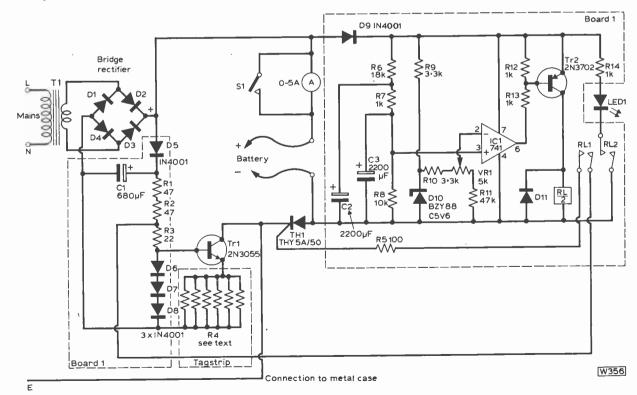


Fig. 1. The complete circuit diagram of the charger. Although a single PCB is used the components mounted on it have been split to enable normal progression through the circuit.

Having determined the charging rate and the end voltage the unit will regulate its on/off cycle quite automatically, providing a rapid charge initially but not tailing off quickly (as a normal charger) when the battery voltage approaches its charged state.

CIRCUIT OPERATION

Fig. 1 shows the complete circuit for the charger.

The mains are stepped down by the transformer T1 and the low voltage output is bridge rectified by diodes D1 to D4 to provide the pulsating DC required for charging.

The additional circuitry of D5 and C1 provides a smoothed DC supply for the base of Tr1. This potential, and the base current, of Tr1 is controlled by R1, R2, D6, D7 and D8.

The charging current has to flow through Trl and, with its base stabilised, the current is limited by the emitter resistor R4. The charging current must also flow through the thyristor Th1 and it is this device which switches the charging current on and off.

★ components list

Resistors R1, R2	47Ω, 5W, Wire wound
R3	22Ω, 5W, Wire wound
R4	see text
R5	100Ω 5% ±W
R6	18kΩ 5% ‡W
R7	1kΩ 5% ±W
R8	10kΩ 5% ‡W
R9, R10	3-3k(2 5% 1W
R11	47kΩ 5% 1 W
R12, R13, R14	1kΩ 5% 1 W
VR1	5kΩ lin. horiz. preset
Capacitors	
C1	680 #F or 1000 #F. 35V
C2, C3	2.200 µF. 16V

D1-D4	Silicon rectifier, 5A, 50V
	individual units or single bridge
D5-D9, D11	Silicon rectifier 1A (1N4001)
D10	Zener diode 5.6V, 400mA
LED1	Any light emitting diode

Miscellaneous

T1, transformer, 16V and 18V tapped secondary at 4A. PCB, Readers PCB Service. RL1, relay, 12V coil, 2 N/O contact sets. S1, switch, S.P.S.T. 5A rated. Ammeter, 0-5A. Box (size will depend on meter and transformer used). Double tag strip, 10 tags each side. DIL socket, 8 pin. Terminals. Battery clips. Grommets.

Note 1

Most of the components used can be obtained from local sources. Minimum requirements have been given and devices better than these can be used. Most mail order companies will supply against the specifications given.

Note 2

Suitable transformers are available from Everest Instruments Ltd., 34 Shakespeare Street, Nottingham and Barrie Electronics, 3 The Minories, London EC3N 1BJ

Time (mins)	Battery Voltage	Current (amps)
start	9	3.4
5	- 11	3.1
30	12	3.0
110	12	2.9
140	12	2.8
200	12	2.8
440	13	2.8
530	14	2.7
590	14	2.8

Table 1. The charging rate and time to bring a flat battery to fully charged. The current and voltage figures have been rounded off. Note that although the charging rate does fall-off it is effectively level over a significant portion of the charging time.

The heart of the thyristor charging circuit is the operational amplifier IC1, which is acting as an under voltage switch. The negative input, pin 2 of IC1, is fixed by tapping across VR1 which is, in turn, regulated against supply variations by zener diode D10. The sampling voltage on pin 3 of IC1 is a fixed percentage of the battery voltage and is determined by the potential divider R6, R7 and R8. Capacitors C2 and C3 slow down the "hunting" action of IC1 when the battery is in the fully charged state.

With a discharged battery connected the sampling voltage will be below the reference voltage and IC1 will cause Tr2 to conduct. The supply for the circuit is obtained from the battery itself when first switched on. When Tr2 conducts, the relay RL1 is energised and a DC current is fed to the gate of the thyristor. The anode of the thyristor is connected to the negative side of the bridge rectifier and the cathode is fed, via the battery, with the pulsating DC from the positive of the bridge. Therefore the thyristor can pass the charging current. Transistor Tr1 already has base current and now has a collector voltage so it too can conduct. The battery is therefore being charged.

When the battery reaches its fully charged voltage, usually set at about 14.3V, the positive input at pin 3 of IC1 exceeds that at pin 2 and IC1 switches off transistor Tr2. In turn, Tr2 de-energises the relay which removes the gate supply to TH1. With no gate current, TH1 will switch off at the next half cycle trough and will not switch on again.

The removal of the charging voltage permits the battery to drop from its "voltage on charge" to its "voltage off charge" and the cycle is repeated. With-out the capacitors C2 and C3 the cycle is governed by the relay energise and de-energise times and the relay chatters.

PRACTICAL CONSIDERATIONS

Before obtaining any components it is necessary to decide on the charging current required. For most home garages about 3A is sufficient. Table 1 shows the time taken for a nominal 3A unit to bring a very flat battery up to its fully charged condition. Although the unit is a constant current design the current does decrease as the battery voltage in- creases but not by a significant amount. The author settled for 3A and the components list is based on that current.

Having chosen the charging current the choice of components can be made bearing the following points in mind.

The transformer should be double wound and have a secondary of 16V to 18V, if higher currents are required then voltages above 18 may be required. Since the secondary is bridge rectified it doesn't require a centre tap.

The rectifier bridge can be an encapsulated unit or separate diodes. Silicon rectifiers are preferred because of space and reliability but it may be possible to pick up a suitable selenium stack at a local shop. Although in a bridge circuit each diode supplies only half the rectified current it is better to use devices with a generous safety margin, say a 5A device for a 3A or 4A charger.

Trl can be a 2N3055 for currents up to 15A but will need a heat sink, (if a metal box is being used, bolt it onto the side). By earthing the collector, as in Fig. 1, there is no need to use insulating washers.

R4 may be dissipating several watts at the higher charging currents and voltages. To assist in setting and adjusting the current several resistors in parallel have been used. The values and numbers will depend on what is available locally. Table 2 gives the currents obtained against various values of emitter resistance. Since the prototype used 2Ω units the table is based on numbers of these in parallel.

Number of 10Ω	Charging	Charing
resistors in	current at	current at
parallei	16V setting	18V setting
(R4)	(amps)	(amps)
10	2-8	3.0
8	2.5	28
5	1.7	2-0
2	not measured	1-1

Table 2. The charging rates against Tr1 emitter resistance and supply voltage. Note that the charging current does not have a linear relationship with the supply voltage. Higher voltages were not tried on the prototype and the requirements for higher currents cannot be specified.

Current setting can be achieved by adding or removing resistors and, subject to the ratings of the rectifiers, transformer and thyristor not being exceeded, no components need to be changed.

Current levels could be changed by switching the resistors in or out but the switch must be rated at several amps. No details of switching are given nor is a switch called for in the components list.

The thyristor passes the full charging current and for a 3 to 4A charger should be rated at 5A minimum. The applied voltage is low and the lowest standard working voltage of 50V is ample. Similarly there is a large gate current available and the device doesn't need to be a sensitive one.

The relay needs two pairs of contacts, both normally open, one pair for the thyristor gate and the other pair for the indicator. When an open relay is used together with a large metal box for the unit it will probably make a sufficiently loud noise to enable the indicator to be dispensed with. In this instance, of course, a single pole relay will suffice.

The relay noise and/or the indicator lamp coupled with the reverse polarity protection built into the unit makes it possible to connect the battery to the charger in complete darkness. If the polarity is wrong nothing happens. No noise, no light and, more importantly, no damage. On reversing the connections at the battery or at the charger the system should operate correctly. The relay resistance should be at least 100 ohms and the transistor Tr2 should be capable of passing at least 200mA collector current.

CONSTRUCTION

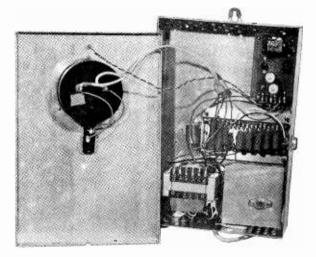
The small electronic components, which will be common irrespective of the charging current selected, are mounted on a single P.C.B. The wiring pattern for this board is shown as Fig. 2 whilst the component positions and orientations are shown as Fig. 3.

The only large component common to all charging currents is the relay. This is also mounted on the P.C.B. but, to leave the choice of type as great as possible, provision is made to make connections between it and the board by flying leads. Some form of mechanical fixing will be required but the type and location will depend on the relay selected.

Although a metal box is not essential for containing the circuitry it has many advantages, both mechanical and electrical.

The power transistor, Tr1, will require a heat sink approximately 3_{12}^{1} inches square for a 3 or 4A charger but could be bolted directly onto the side of a metal box. Since the design earths the collector of this transistor there is no need to use insulating washers.

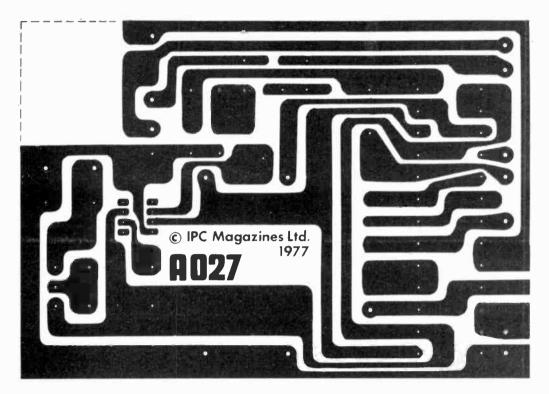
Whether the box is metal or not, heat sinks will be required for the thyristor and the bridge rectifier. These can be of 16 gauge aluminium approximately 3^{1}_{2} inches square. To save space they can be folded into channel sections but must be mounted vertically to ensure an adequate air flow. The thyristor must be fitted with insulating washers and if individual stud mounting diodes are used they too must have insulating washers.



A photograph of the inside of the prototype. This unit did not use a PCB and several of the components now mounted on the board were mounted on the tag strip.

The resistors making up R4 were mounted on tagboard. This made changes easy and would assist switching if required. Although 2Ω units were used the individual components can be of any low value.

Although the prototype didn't get more than just warm, holes were drilled in the ends of the case to enable an air flow when fixed to the wall. Several



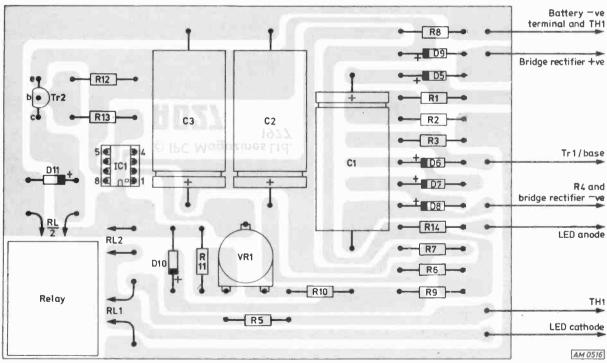


Fig. 2. The upper drawing gives the wiring pattern of the PCB designed for this charger, drawn full size. The individual tracks are capable of carrying well over the 3A used. The lower drawing shows the location and orientation of the components.

small holes are better than a few larger ones unless a screen is used to prevent items falling in.

When the battery has reached its "fully-charged" state the relay will de-energise. Because of the differences in "on-charge" and "off-charge" voltages the relay will pull in and drop out all the while the charger is connected. The rate will depend on the

charging current set but will be of the order of 10 seconds. The meter will try to follow this change from full charge to no charge and will seem to fluctuate. To minimise possible damage to the movement a switch has been incorporated and the meter can be short circuited. The relay noise and the indicator will show that the charger is still working.



Order in comfort

Regular readers of PW, will I'm sure, have seen the Maplin advertisements regularly carried on the back cover of this magazine. Now, to make the job of ordering components an easy, accurate and enjoyable pastime, Maplin have come up with a brand new catalogue containing hundreds of photographs and illustrations. Measuring 285 \times 205 \times 10mm, this latest offering contains 216 pages packed with details of components, circuit ideas, diagrams & PCB's, kits for synthesiser, organs, amplifiers, and sections containing tools, books and cabinet hardware. In fact just about everything the DIY electronics enthusiast requires.

The catalogue comes complete with up-to-date price list, order form, postage paid envelope, and money back guarantee if returned within 14 days of purchase. The price incidentally is 50p.

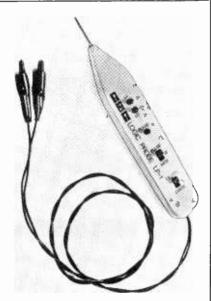
For those interested, contact

Probe for faults

Electronic enthusiasts are well aware of the ever increasing use that is made of DTL, TTL/C-MOS devices in almost every type of circuit that not so very long ago would have necessitated the use of hundreds of transistors. Unlike transistors though, logic IC's are generally more difficult to test in circuit, especially C-MOS devices which require extreme caution, if they are to survive the rigours of testing.

A logic test probe is the answer, and the latest to appear in the UK market is sold by Rastra Electronics Ltd, who have launched the LP-1 logic probe. A few of the parameters of this probe are a detectable pulse width of 50nS, a maximum input frequency of 10MHz and a power requirement of from 5 to 36V. LED's are used to indicate logic '1' (2.25V \pm 0.015V 70%Vcc) or logic '0' (0.8V \pm 0.01V 30% Vcc) and pulse, which blinks on for $\frac{1}{3}S$ transients. Price is £28.60 plus VAT. Maplin Electronic Supplies, PO Box 3, Rayleigh, Essex. Tel: 0702 715155 or call at Maplin's shop 284, London Road, Westcliff-on-Sea, Essex. Tel: 0702 47379.





Rastra Electronics Ltd., 275-281 King Street, Hammersmith, London W6 Tel: 01-748 3143

I see new chips

Three new devices that have recently come onto the market are discussed below and are available from Rastra Electronics, Distronics, and SGS-ATES, respectively:—

The first is a precision waveform generator/voltage controlled oscillator, and designated the ICL8038-CCPD. The IC is said to be capable of producing sine, square, triangular, sawtooth and pulse waveforms at frequencies ranging from less than 1/1000Hz to more than 1MHz. FM and sweeping can be accomplished with an external voltage, and it is claimed to be highly stable over the temperature range of 0°C to 70°C. Package is a standard 14 pin DIL. Further information from Rastra Electronics, 275-281 King Street, Hammersmith, London, W6. Tel: 01-748 3143.

Continuing with oscillators, Distronic have released details of their crystal controlled oscillators, which combine crystal techniques with hybrid integrated circuit technology.

The units are claimed to have the high stability temperature and aging characteristics associated with AT-cut quartz crystals, and are available for driving multiple TTL or C-MOS loads. For TTL types, the available frequency range is from less than 0.25Hz up to 30MHz with an input voltage of 5V DC.

The frequency range of the C-MOS oscillators is from 300Hz up to 10MHz for TO-5 package types and from 0.0002Hz up to 10MHz in 14 pin DIL packages. Distronic Limited, 50-51 Burnt Mill, Elizabeth Way, Harlow, Essex. Tel: Harlow 32947.

The new SGS-ATES chip is a 5 terminal voltage regulator-the L200. Output voltage is adjustable between 3 and 30V and output current adjustable from 0 to more than 1.8A. Thermal overload and short circuit protection are also incorporated, as is second breakdown and overload protection which enables the L200 to withstand spikes of up to 60V. Load regulation is said to be 0.1% of Vout, while ripple rejection is better than 70dB. Other applications for this device are precision current limitation and switching regulation in which neither external transistors nor additional protection circuitry is required.

Further information on this product from SGS-ATES (UK) Ltd., Walton Street, Aylesbury, Bucks. Tel: 0296 5977.

Disco Separates

Rather than 'Cart around' a heavy disco comprising of decks and amplifiers combined in a single cabinet, it is often advantageous to have separate units. The new RTVC 70W and 100W RMS mono disco amplifiers fit into this category, each providing a full range of mixing facilities and an input for a separate slave mixer amplifier. Both units are fully protected against speaker malfunction and against any short circuit conditions. Maximum power is developed into 40hms.

Inputs 'Deck 1' and 'Deck 2' are for ceramic cartridges with outputs in the order of 200mV, while the 'Mic' input is a dual sensitivity type, accepting either 1mV or 10mV outputs. Microphones may be of either high or low impedance. Other inputs provided are 'Tape in' (300mV into 50k ohms), 'Tape out' (150mV at 100k ohm) and 'Slave in', 'Slave out', both at 150mV at 5k ohm.

PFL—pre-fade listening is also incorporated on each input regardless of the level, and is controlled by a separate rotary control.

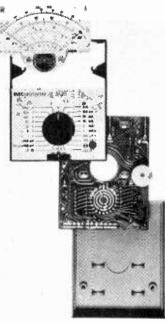
Along with other RTVC products, these Disco Amplifiers are very competitively priced, with the 70W model selling for £49.00 plus £3.00 p&p and the higher powered 100W model carrying a price tag of £65.00 plus £4.00 p&p.—RTVC, (Dept. PW), 21 High Street, Acton, London, W3 6NG and 323 Edgware Road, London W2.



4-piece meter

One of the most frustrating factors in owning any instrument is that of servicing and calibration. Alcon Instruments claim that they have solved this problem with the introduction of their new pocket-size range of multimeters from Miselco. These instruments have been designed for repair in a simple and easy manner using factory matched replacement modules which, by virtue of their design, obviate the need to recalibrate each time.

The Miselco Tester range of four Multimeters cover between them a high current model capable of measuring up to 30A; a model with a high sensitivity figure of $1M\Omega/V$; a general purpose model with a sensitivity of $20k\Omega/V$ and finally one intended for electronic work with a sensitivity of $50k\Omega/V$ The latter model has claimed accuracy figures of 2.5% on DC and 3.5% on AC,



while all the other models claim 2% on DC and 2.5% on AC.

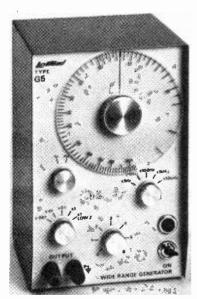
Each model comprises two case parts, a movement and a PCB Any one module may therefore be replaced by the user, using only a screwdriver to complete the job.

Alcon Instruments Ltd, (Dept. PW) 19 Mulberry Walk, London SW3 6DZ.

put voltage of SV RMS can be reduced to less than 1mV by means of a continuously variable control and the switched attenuator. Size is 130 x 130 x 121mm and is priced at £58.70.

Linstead Electronic Instruments, (Dept. PW), Roslyn Works, Roslyn Rd, London N15 5JB. Tel: 01-802 5144.

Everymans Sig. Gen.



Practical Wireless, May 1977

Test equipment these days tends to be so expensive, that only the largest of service centres can afford to keep up with new products. This of course leaves the average enthusiast completely 'high and dry' when it comes to the latest techniques in test equipment.

To overcome this apparently insurmountable problem, Linstead have launched a signal generator that is one of the cheapest on the market today, but without sacrificing quality or facilities. Called the Linstead G5, signals can be of either sine or square wave, and range in frequency from 10Hz up to 1MHz in steps of 1Hz on the lower range. The output is fed from a power amplifier which allows output impedances from 600 ohms down to 5 ohms.

Claimed accuracy figures are 2% +1Hz with sine wave distortion less than $\frac{1}{2}\%$ harmonic content. The out-



Designers-

please note

I am one of the unfortunates so roundly condemned by Paul Heath of Wolverhampton on your letters page in the February issue. Unfortunately my QTH is quite useless for VHF transmitting, so my activity is therefore confined to low-power portable working using FM. If it were not for repeaters such as MP and RF, I would have no contacts at all. I rarely hear any signals on the simplex channels.

Working through repeaters is a great-boon for mobiles also, handicapped by low power and whip aerials. And if we all incur the wrath of the critics for not listening on the input channels, that is because of a shortage of receiver designs. P.W. has published some splendid articles on 2-metre converters, and on hi-fi equipment for the FM broadcast bands. But *never* have I seen a design for a straightforward FM receiver for 2-metre work.

May I suggest a possible specification? (a) single conversion receiver with tunable front-end, using a variable capacitance diode; (b) integrated circuit (e.g. CA 3089E) for the complete IF strip, possibly including the cheap Toyocom crystal filter to avoid the headache of alignment problems; and (c) a small integrated circuit for the audio side. such as the LM 380. This receiver would work from the 12V vehicle battery or from a small mains power pack. I'm sure that there would be a great interest in it on the part of SWLs, many of whom enjoy listening to the repeaters but are handicapped through the use of the inefficient method of "slope detection".

PW has many gifted contri-

butors: surely one of them could come up with an effective design along the lines indicated. And then you could follow it up with a design for a converter for the 70cm band, to work into the PW FM receiver! F. Ness GD3ESV (Douglas, IoM).

Hot tip

Re the stripping of Litz wire (Letter H. Pruim) in the February issue. I am now retired after a lifetime in many branches of work using all sorts of wires, and for years I have used a simple method of stripping Litz that needs only a small meths lamp with wick and shallow tin lid.

Just put about a ¹4in of meths in the lid and put it by the lamp. First hold the end to be stripped in meths flame till just red hot. Then quickly dip the hot end into the meths, presto—perfectly clean strands just right for soldering. Not much knack required, but a few trys will soon prove the idea. George Comber (Tadley).

High power interference

Democracy, according to the Radio Society of Great Britain requires that minority groups be given absolute priority over other amateurs—or so it seems on the 2m band. I will not challenge Dr. Allaway's remarks in your February issue about minority groups having a right to exist, but they must not be allowed to take up unduly large portions of any band, in a way that absolutely prevents anyone else using that band segment.

I am referring to the Oscar command station which is located at Surrey and requires about 100 kHz of the 2m band. The power which is radiated (believed a few kW E.R.P.) completely obliterates any but the most local legal-limit transmissions. The transmission is in the "all-modes" part of the band, which should be shared by all and not monopolised by a minority service which already has its own "space" segment elsewhere.

The RSGB required the greatest

persuasion even to publicise the source of the nuisance. They will now have nothing more to do with the matter—I cannot even get them to tell me when to avoid the frequency! The transmission begins without warning about once every two hours or so. It has caused interference as far north as Liverpool.

"RSGB represents all UK radio amateurs." It says so on my membership card. The blatant white-wash over the mistake in licensing this wide-band transmission in an Amateur band seems to suggest that the statement is in some way at fault.

How can I trust a society that performs in this manner to safeguard my interests at WARC in 1979? It is the same society that had a 6-month option on a very costly investment, and although promising "informed discussion" on the wisdom of the purchase and suggesting that they would wait to hear any views that informed members may have on the subject, they abruptly bought the machine in question with most of the 6-month option period still left to run.

Yet the RSGB is probably the only body that can make representations to the delegation from this country that will vote at WARC.

RSGB, you are in a position of trust. You must be above suspicion, in all your claims. Please realise your responsibilities.—G. L. Manning, G8JBH (Edgware).

More 'scope

Just a small but important observation I thought we ought to share. Many useful projects have been featured in PW associated with oscilloscopes including the "bolt on" goodies and details on operation and fault-finding techniques. However, many of us, and specifically myself, haven't got the basic and necessary item, i.e. **an oscilloscope.** The last featured scope, correct me if I'm wrong, was August 1973 "PW Student".

Could it be that 1977 is going to give us a super new scope project? Well, that's it, except one specification. Please! a minimum 4 inch tube. G. Starkey (Warwickshire).

The Editor will be pleased to hear from potential contributors on the above projects.

Practical Wireless, May 1977

COMPACT 2m BEAM AERIALS

THE previous article "Vertical Aerials for 144MHz" (Practical Wireless July 1976) dealt with omnidirectional types such as the ground plane and vertical colinear which are convenient for transmitting or receiving in any direction but have relatively little gain over a dipole. For example, a 4-element colinear has but 4.3dB gain and is a sizeable aerial, some 14ft or so long. Beams are the only way of obtaining relatively high directivity and gain and there are many to choose from with the Yagi or parasitic array being the most popular.

However, fairly small but efficient beam aerials for 144MHz are not difficult to make and gain as high as 10dB over a dipole is possible without resorting to large numbers of elements as are necessary for Yagi or colinear arrays. The corner reflector aerial, which employs only one driven element, may be an exception and although a high gain, around 12dB or more, is possible, the reflector itself is of somewhat unwieldly dimensions.

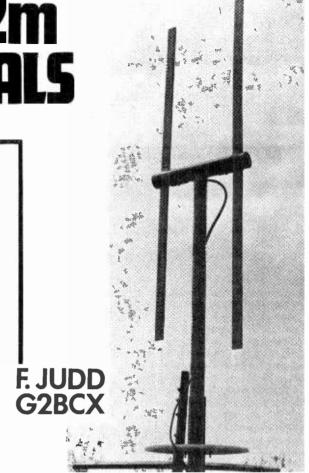
TWO DRIVEN ELEMENTS

Small beams with a useful degree of directivity and gain may be derived from "end-fire" or "broadside" arrays consisting of two driven radiators spaced a specific distance apart with the current in each phased according to the directivity required. In the case of the "end-fire" array the pattern has zero radiation broadside (at right angles) to the plane of the array and maximum end-fire radiation when the spacing between the elements is a half-wavelength or less and the currents in the elements are equal in amplitude but in opposite phase i.e. 180° phase difference.

. The configuration of this array is shown in Fig. 1 in which the spacing "d" may be from ${}^{1}{}_{8}\lambda$ to ${}^{5}{}_{8}\lambda$ to obtain the requisite horizontal plane end-fire radiation pattern when the array is vertical as in Fig. 2(a) and looking down on the ends of the elements. Radiation in the vertical plane i.e., with reference to an angle to the ground, will be as shown in Fig. 2(b).

To obtain "broadside" radiation, at right angles to the array, the spacing between the elements is usually a half-wavelength and the elements fed in phase. The available directivity/gain is greatest when the elements are spaced ${}^{1}_{2}\lambda$ or ${}^{5}_{8}\lambda$ but not closer. The end-fire array is of greater interest since considerable gain can be obtained with closer spacing between the elements thus allowing for a more compact aerial plus uni-directional or directional radiation pattern as required. Such an aerial may be operated horizontally (for horizontal polarization) or vertically (for vertical polarization).

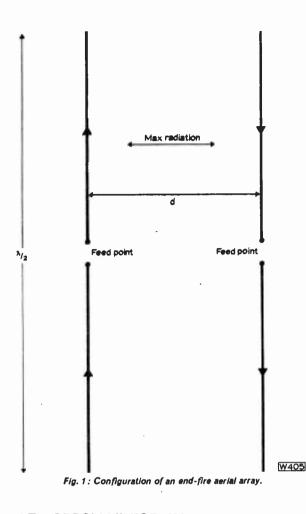
Greatest bi-directional gain is obtained when two driven half-wave elements are spaced 0.125λ apart



and fed out of phase as in Fig. 3. The gain is about 4dB with reference to a half-wave dipole. However, for 144MHz operation the high feed point impedance presents a problem as it is not suitable for direct connection to commonly used 50Ω co-axial cable although this can be overcome by the use of a matching stub.

Since beam aerials have to be rotated for maximum radiation in a desired direction there is little point in using a bi-directional array. Better to use a uni-directional system and obtain a little more gain. This can be achieved with a two-element close spaced array in which one element is fed 135° out of phase with respect to the other. The radiation pattern then becomes a cardioid as shown in Fig. 4 and remains virtually the same whether the aerial is used vertically or horizontally.

An aerial based on this arrangement, to provide even more gain, was designed by the writer some 26 years ago and is known as the "ZL Special" although it was originally intended for HF bands operation. It employs two driven elements fed 135° anti-phase but the element lengths are such that one operates as a driven element/director whilst the other behaves as a driven element/reflector. The radiation pattern is cardioid as in Fig. 4 and the gain a little over 6dB with reference to a half-wave dipole.



"ZL SPECIAL" FOR 2M

This beam is identical to the original model except that an alternative feed method to suit 50Ω co-axial cable was necessary and the element lengths are appropriate for 2m band operation. It is easy to construct and the elements and phasing line may be made from copper wire, copper tube or even 300Ω ribbon feeder. Quite a large number of this 2m version have been constructed and tested and being a very compact aerial it has proved ideal for use indoors since it is not generally affected by the proximity of walls and even conductive structures, providing these are not too close.

This beam has a broadband characteristic and properly constructed and adjusted should exhibit around a 1:1 SWR over the whole 2m band. Constructional details are shown in Fig. 5 and providing the phasing line length and the element spacing and lengths are adhered to, the materials for these may be as mentioned above. If 300Ω ribbon feeder is used the distance between conductors will of course be that of the feeder itself.

The small air-spaced 20pF variable capacitor is necessary to achieve a correct match to 50Ω co-ax cable and is adjusted for minimum SWR at 145 MHz. Not more than about 10pF of this capacitance should be necessary to achieve this. Do not use mica or ceramic type trimmer capacitors. It is important to curve the feed line away and down as shown in the inset in Fig. 5 when the aerial is operated vertically

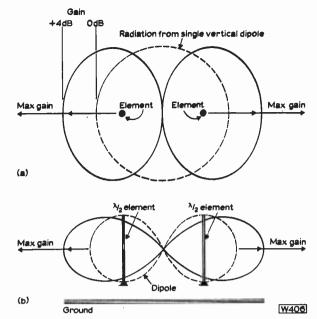


Fig. 2: Radiation patterns and gain of an end-fire array compared with a dipole.

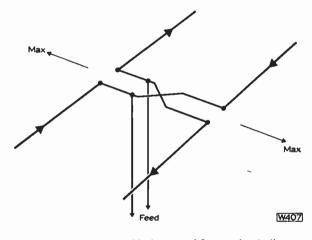


Fig. 3 : One method of feeding an end-fire array (see text).

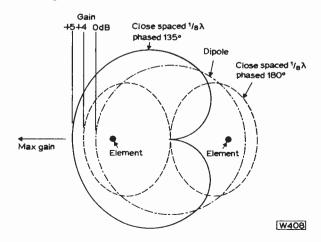
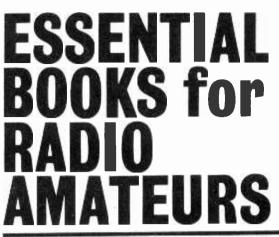


Fig. 4: Gain and radiation pattern of a close spaced end-fire array with currents 135° anti-phase.

Practical Wireless, May 1977

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SERVICING FEATURES

Amongst the servicing features next month will be an article by John Coombes on the GEC single-standard monochrome chassis and advice on the French (EMO, Eurovox and Eurosonic) colour receivers that were imported into the UK during the colour boom period.

THE COLOUR SUBCARRIER PARADOX

As we all know, suppressed subcarrier transmission is used for the colour signal in the PAL and NTSC systems. Yet a 4.43MHz signal is still present. Do you know why, or what effect suppressing the subcarrier has on the composition of the transmitted signal, and how this affects modulation? These matters are basic to understanding colour television but are seldom clearly explained. E. J. Hoare remedies this situation and in doing so brings home the full subtleties of the PAL system

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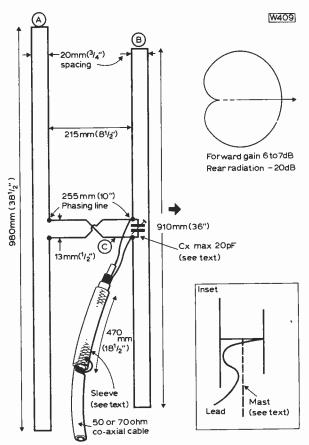


Fig. 5: Details for the construction of a ZL Special aerial for 2 metres.

and the matching capacitor should be adjusted only when the feeder has been secured. The short stub mast used to support the aerial must be of insulating material, wood or plastic tube and extended to a few inches below the bottom ends of the elements when it may be coupled to a metal mast. The gain remains the same when the aerial is used horizontally but the cardioid pattern becomes a little narrower to the sides. Note the unbalance-to-balanced co-axial connection using a ${}^{1}_{4\lambda}$ sleeve which is bonded to the main co-axial braid only where shown, at point X in Fig. 5. The end near the feed point is not connected to anything. For use outdoors the feed connections and tuning capacitor must of course be protected from weather as must the elements and for this a small plastic electrical junction box could be used. The elements and supporting frame (this must be wood or plastic tube) can be given two or three coats of polyurethane varnish.

This aerial is used by the writer on a cabin cruiser (G2BCX/MM) at a height of only 9 feet above the water and so far 10 European countries as well as many UK DX stations have been worked. The photo shows the aerial enclosed entirely in plastic "plumbing" tube for complete protection from weather and salt water. The aerial is also ideal for portable work as it will fit comfortably in the boot of a car.

Since the basic ZL is a driven array and the radiation pattern uni-directional it has been found possible to add parasitic director elements to obtain an increase in forward gain. One director mounted

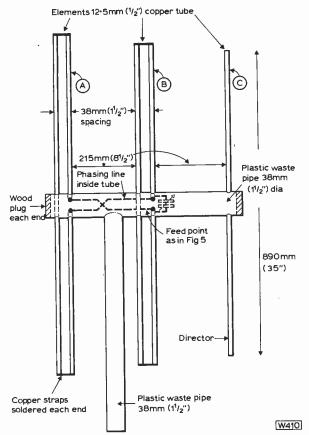


Fig. 6: Construction of a three element ZL Special beam.

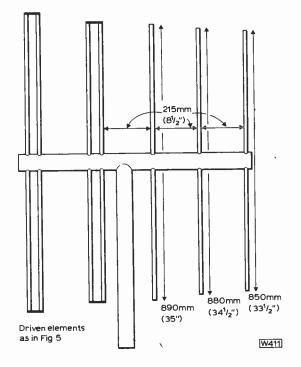
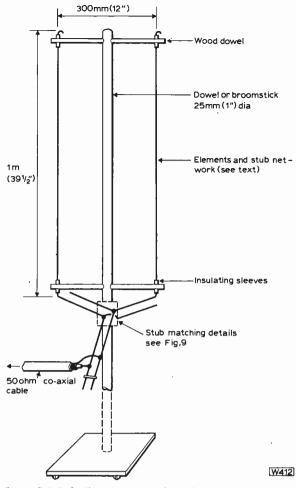
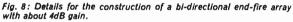


Fig. 7: Details for the director lengths and spacing for a five element ZL Special. Other constructional details as in Fig. 6.





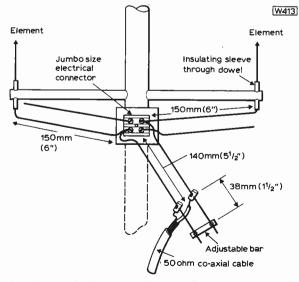


Fig. 9 : Details of the feed system for the end-fire beam shown in Fig. 8.

 0.12λ in front of the main driven element (B in Fig. 5), increases the forward gain by slightly over 2dB thus providing a total gain of 8dB which is quite considerable for a beam measuring only about

50cm (20in.) from front to rear. The addition of three directors, still making the beam only about 1 2m (40in.) high, yields a forward gain of a little over 10dB which is comparable with that of an 8-element Yagi having a physical length of about 2.8m (110in.), or twice the length.

The construction of a ZL with directors is much the same as for the basic aerial shown in Fig. 5 except that the element support is extended for mounting the directors. This must be insulating material, such as wood or plastic tube, and one suggested method of construction, particularly suitable for outdoor use, is given in Fig. 6. The main support which consists of the boom and stub mast is made of plastic pipe. The stub mast is fitted into the boom either by cutting a hole in the boom and gluing the mast in with Araldite, or by using a pipe "T" piece to make the join. The copper tube elements are fitted tightly into holes through the boom section and secured by screws through the side of the pipe. The phasing line which may be two copper wires (12 to 16 SWG) are inside the tube as is the small capacitor. To ensure that the phasing lines do not touch, they may be insulated with sleeving.

This 3-element ZL Special has been tested over a long period and operated both vertically and horizontally with considerable Continental DX worked. The half-power beamwidth (3dB) is about 90° and radiation directly from the rear is about 25dB down although the overall radiation pattern is still cardioid but narrower than that from the 2element ZL Special.

The addition of two more directors, making a total of three, and the aerial therefore a 5-element version, will provide a gain of about 10dB. The aerial is still not very large and may of course be operated vertically or horizontally. Construction is the same as for the 3-element version except for the three extra directors cut to length and spaced as shown in Fig. 7. Note that the amount of capacitance required across the 50Ω feed to effect a good match may be much less or even nil with the directors in use.

VERTICAL END-FIRE ARRAY

This is relatively easy to construct and will provide about 4dB gain over a dipole and is bi-directional. It needs only to be turned through 90° to achieve all round coverage and being compact could prove useful to flat dwellers unable to put up an aerial outside. It consists of two half-wave radiators fed 180° anti-phase to produce the radiation pattern as in Fig. 2a. It may be assembled on a wooden frame and the elements and stubs can be made from copper wire as shown in Fig. 8. It could also be set up to rotate on a floor stand, as the extension of the diagram illustrates, making it suitable for indoor use. Details of the matching stub and feed are shown with more detail in Fig. 9. Adjustment for minimum SWR is carried out by sliding the shorting bar along the stub in conjunction with moving the tapping point of the 50 Ω co-ax feeder. The approximate positions for these are given in Fig. 9 and very little further adjustment should be needed to obtain an SWR approaching 1:1 over the whole 2m band. Used in the writer's workshop at about 10ft. above the ground, this aerial could receive a substantial signal from the GB3PI repeater nearly 40 miles away. p_{W}

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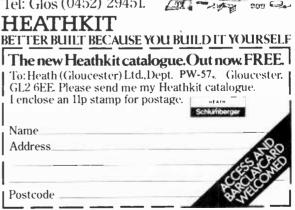
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by Eric Dowdeswell G4AR

APPY man this month is Neil Braeman of Southwater, near Horsham, and regular contributor to this column. He has passed his RAE and he is now G4FUP. Congratulations Neil, I know the bit of effort required to get down to the necessary studying will be found to have been worthwhile, for very many years to come. He has sold his bike to get funds to buy a better receiver and intends to make his own transmitter, a policy which I heartily endorse!

When John Taylor of Cheadle Hulme found the cost of a decent receiver a bit beyond his means he dragged out an old R107 that hadn't worked since 1953! A simple signal generator and multimeter and it was soon going again. John is thinking of building a converter for the higher frequency bands which is a very sound arrangement. In Stocksfield, Northumberland Simon Robinson is just starting to listen to the SW bands and is thinking of getting a surplus CR100 and 'hotting it up'. Well, Simon, unless you have the requisite test gear and a bit of experience I should tread warily. Make one change at a time, re-aligning if necessary, before doing anything else. Personally speaking, I would get it going properly on the LF ranges and use a convertor, as previously mentioned.

A note from Jim Martin A7089, Secretary of the newly-formed Edinburgh and District ARC, tells me that a group of about 20 members meets every Tuesday at 1930 at the club's rooms situated at the City Observatory, Catton Hill, Edinburgh. One could say that things are looking up! The club is active from 2m to 160m with calls GM4AJV/A and GM4BWT/A until they get their own call. Interested? then write to Jim Martin at 22 Ross Gardens, Edinburgh EH9 3BR. Still in GM-land, the Aberdeen ARS got all excited recently with a visit from the BBC involving interviews with several members, later broadcast on Radio Aberdeen. The producer is interested enough to suggest a monthly spot on Radio Scotland dedicated to amateur matters. In the meantime more mundane meetings take place at the hall to the rear of 91 Crown Street, Aberdeen. On 8 April there's a demonstration of an automatic CW reader and a try out of a new Jap receiver that covers LW to 500MHz! On 15 April an RSGB Tape Lecture will be illustrated with slides while 22 April sees RSGB Council member GM3ZBE visiting the club to talk of Society affairs. Secretary Stan Sutherland awaits your enquiries at 67 Greenfern Road, Aberdeen or ring 691716.

Practical Wireless, May 1977

Back to regular correspondent Robin Bayley A9203 who is now settled at Kemberton, near Shifnal, Shropshire. He has been assessing his new QTH from the DX aspect from 15 to 80m with the aid of his EC10. With HK0's on 80, 40 and 15m and KH6 on 20m things look promising. Andrew Work A9091 in Beverley, E. Yorks, changed his 66ft wire round the loft to a straight wire outside and naturally found a big improvement in results including VP2 and VP5 on 80m. In Wallasey E. J. Shaspe has progressed through a couple of receiver kits to an Eddystone 670A which he found going cheap. He's anxious to beg or borrow a handbook on the same so if you can help write to 36 Warren Drive, Wallasev, Merseyside L45 0JS.

Zone 23 is a very rare place and many aspirants for the Worked All Zones award have had long waits for a QSL from there. B. Harrison in Hastings heard JT1KAA from Mongolia which is a good start to getting a card. KC6 was another good catch in the East Caroline Islands in the Pacific. Steve Budd A8713 also on the sunny south coast, in Worthing, has been logging stuff on the downlink from Oscar 6 10m and the 2m downlink from Oscar 7. However he still did not desert the HF bands. Steve Cottis A8961 seeks the DX from Harrogate and has QSL's from HP7, VQ9 and VR8 to prove he has been successful. I'd hesitate to attribute the improved conditions on 15m to a general rise in sunspot activity. Steve. Rather it is the usual improvement as we move towards the summer in the northern hemisphere.

David Peck BRS37621 has done us proud this month from Cambridge with logs for SSB and RTTY somewhat restricted by a move with all the gear from the house to a shed in the garden!! David started with a Creed 7B printer plus 'RTTY the Easy Way' built the terminal unit and he was away! The receiver is a Yaesu FRG7 fed from a 135ft wire. D. Anderson BRS36591 has not written for a while from Brookwood, Surrey, but makes up this month with a good log. Normally he tapes his DX but at the moment the recorder is u/s. Dennis takes the RAE next December and in the meantime G8LVB is helping him with the necessary studies.

Now a note for all! The Annual Convention and Exhibition of the Northern Radio Societies will take place at Belle Vue, Manchester on Sunday 24 April. See the box in the News feature of this issue for further information.

Log Extracts

J. Taylor:- 20m KG4DHD VE8SO 7X2LTA

R. Bayley: 80m AP2AD EA9CR FP8DX HK0COP JA1JRK KH6PP PY7PO 40m CT2AP EA9CF HKOCAT 20m FP8DH KH6PA TG9TL VP8AA 15m EA6DA FP8DA HK0KOC TG9TY ZL3AK

A. Work:- 80m VP2DAC VP5LDH

B. Harrison:— 20m FG7AK FM7WE FY7YM HCISC HK0BKX KG4PF PJ2FR SV0WZ (Rhodes) TI2FM VP2EEQ (QSL WA6AHF) ZD7SD 40m HCIDK VP2KF ZS3AW 20m P29PN S79P (Seychelles) 7P8AC 7Z1AB

D. Anderson:- 80m VS6DO 7Y4MD 20m C21PS (Nauru) CP5GR HM0SM KC6KO P29BS VP2AG **ZP6KR**

D. Peck:— 20m KC4AAA KC4USD KL7IEU 15m JR6HJJ WB2CJF/HZ1 20m RTTY IC8POF K9KHI OHONI UK3ACR WA0YDJ/4 YU2CE 9K2EP

N. Braeman:— 80m KP4AST 9Y4NP 20m OE5GML/YK PZ1DR W6QL/VP2A(CW) YS7VI

S. Cottis:— 80m CT2BU KP4EBH 20m JW7FD KL7IEU VR4DX 15m D2ASW FP8DX HP7XJS WB6EWH/VQ9 (Chagos) 3B8CV 5U7AG

S. Budd:— 80m FMTWS JY9CR PJ8KG TI2FJC 6Y5EM 20m D2AFE (Ex-CR6FE) FR7ZL/T (Tromelin) TR8LE W6QL/VP2A YB2CR 8P7GN 15m N4PN (USA) PY0ZAE (QSL PY1CK) 7P8BC Oscar 6 (10m) GJ8EZA VE3EYR Oscar 7 (2m) CT1WW HW6ADB IT9ZDA K4UQ TU2EF VE3FKU W2BXA W9QOO

All SSB except where stated otherwise.



MEDIUM WAVE DX

by Charles Molloy

report from Ralph Newman in Reading mentions an unidentified station on 945kHz between 2300 and midnight which created a 1kHz heterodyne with the Russian on 944kHz. The writer has made a tape of the closedown announcement, which is in English, and although no location was obtained, two frequencies on the medium waves were referred, to 602kHz and 945kHz. The transmission ended with the Nigerian national anthem at 0002. The Western Nigeria Radiovision Service has a 10kW outlet on 602kHz located at Abafon and it seems likely that they now have another on 945kHz. Although the signal is quite strong at times interference makes reception difficult. DXers who use communications receivers and a loop might like to take up the challenge offered by this new broadcaster and try to obtain a more definite identification.

In a letter from Horsham in Sussex, Kes Otley reports hearing CKVO Clarenville in Newfoundland on 710kHz using his Roberts portable radio connected to a long wire aerial and earth. He asks how to use a loop with this receiver. From St. James in Barbados, Paul Griffith poses a similar question when he asks how can he connect a loop to a portable receiver which has no external jacks for earth or aerial. A number of DXers have written on similar lines and the short answer is that a loop is really unsuitable for use with a receiver that has an internal aerial. Even if aerial and earth sockets are fitted to the portable, they are intended for use with a wire aerial and if a loop is connected to them then poor results will be obtained.

The internal aerial in a transistor portable is wound on a ferrite rod or slab and this type of aerial is directional. There are two positions of maximum pick-up at right angles to the rod and two other directions of minimum pick-up which lie along the length of the rod. A loop aerial which is based on the frame aerial used in the early days of radio, has similar properties and if an attempt is made to use both aerials at once then problems will arise. The loop may detune the receiver input causing poor selectivity. If the null (direction of minimum pick-up) of the loop is pointed at an interfering station, there may still be pick-up from that station via the internal aerial, the overall effect being to reduce the directional properties of the loop.

One method of overcoming this problem is to attach the receiver to the centre of the loop in such a way that the nulls of loop and receiver coincide. In most cases this will mean that the receiver will be at right angles to the plane of the loop windings. Coupling between loop and receiver is inductive, so no direct connection should be made between them. The loop and receiver are rotated together to null out interference. This method should provide an interesting field of experiment but at best it is only a makeshift.

The most effective DX tool on the medium waves is a communications receiver and readers will note that most of the DX reported in this column comes from users of this type of equipment. Newcomers to the hobby who are thinking of investing in a receiver should think along the lines of a Realistic DX160, Trio 9R59D/E or the Heathkit or Eddystone range, rather than of portables or hi-fi tuners which do not possess the selectivity or sensitivity required by the serious DXer.

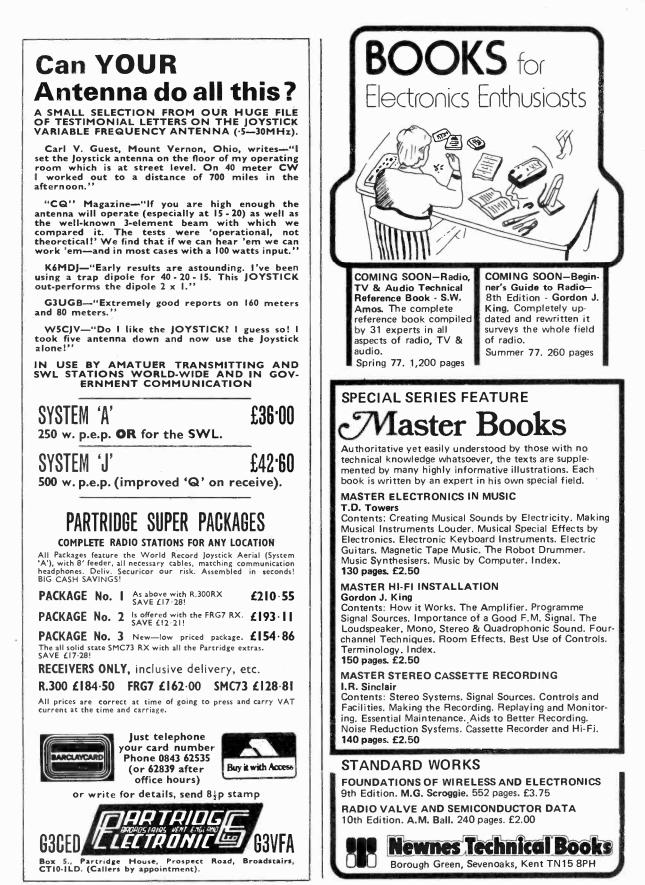
Robin Bayley has moved to a new QTH at Kemberton in Salop and is now the owner of an Eddystone EC10. He reports hearing a DX programme from Thames Valley Radio on 1430kHz during the Golden Days programme which is on the air on Sundays between 2200 and 2300. Another programme of interest to the DXer is the weekly 'Sweden Calling DXers' which is on the air on Tuesdays at 2315.

Latin American medium wave stations come in well in the UK, the long sea path across the Atlantic favouring propagation from this area. Reception from South America is often at its best when DX from North America is poor so it is worthwhile in vestigating Spanish and Portuguese speaking stations when DX from North America is absent. It is unusual if there is no DX at all on the medium waves! John Wildey who uses a Grundig Satellit 2000 in Doncaster says he has positively identified Radio Margarita on 1020kHz but he finds that the fast speaking style of South Americans makes identification difficult. He has heard a number in the 800 to 1000kHz and 1200 to 1400kHz areas.

Latin American countries most frequently heard are Colombia, Venezuela, Argentina and Uruguay where Spanish is the language and Brazil where Portuguese is spoken. Although callsigns are allocated they are seldom used but most broadcasters have names such as Radio Globo, R. Nacional followed by the location, such as Radio el Mundo Buenos Aires (which is on 1070kHz). The 'World Radio and TV Handbook' gives a comprehensive listing of LA stations both by frequency and country together with station addresses. In answer to J. Williams of Frome in Somerset, the WRH is distributed in the UK by Fountain Press and it can be ordered through bookshops.

"Wonders never cease, I got a verification yesterday from Radio Sutatenza, Magangue, in Colombia, HJHN on 960kHz" writes **Harold Emblem** from Mirfield in Yorkshire. Latin American stations are notoriously difficult to verify. Receiving a station is only the first and usually the easiest part of the exercise. They are unpredictable too. The DXer can plug away for years with reports that are unanswered and then suddenly receive a cordial letter with pennant, QSL card and detailed information about the station and its locality. This appears to have happened to Harold who now has details of





the Radio Satatenza network in Colombia. Each station calls itself Radio Satatenza following this with the name of the town. HJCR Medellin is on 590kHz with 100kW, HJCX in Cali is on 700 with 120kW, HJCY in Bogota on 810 with 250kW, HJHN in Magangue on 960 with 120kW and HJGL in Barranquilla on 1010 with 10kW. HJHN in 960 is often logged by DXers in the UK.

More LA DX comes from **Ralph Newman** in Reading with his homebrew receiver and loop which pulled-in Radio Margarita on 1020 and Radio Vision, the 50kW outlet in Maracay, Venezuela on 650kHz. **Roy Patrick** who DX'es in Derby with a Trio 9R59D and loop reports hearing Trans-World Radio Bonaire on 800kHz at 0200. This station is on the island of Bonaire which is part of the Netherlands Antilles which lie off the northern coast of South America. Multi-lingual programming, which includes English is radiated from a 500kW transmitter and is mainly of a religious nature.



SHORT WAVE BROADCASTS by Derek Bell

F IRST we must tie up a loose end from a month or so back. **Tony Cook** had an HAC set that was giving trouble on its band spread section and thus thought it wise to ask, via this column, if his confréres in the DX world could assist. As a result a letter is before me from **Joe Owens** of Dolgellau, North Wales who describes himself as "an old time PW'er from the nineteen thirties". He suggests that it might be worth trying to connect the BS capacitor to the main tuning capacitor via a small series fixed capacitor. If Tony has not got one of these he can twist two equal lengths of thin wire together. Well there seems to be enough there to pass a few happy hours tinkering! I hope that it achieves the required result.

As a footnote Joe says that he has six redundant loudspeakers that need good homes and for the price of the return postage he will pass them on. If I may inject a personal note here it might be nice if they went to school clubs or disabled or pensioner DX'ers.

For once in my life, as the song says, I can be in the forefront and let you have advance news of a station. This rare event is due to the good offices of **Robin Bayley** from Kemberton, Salop. Robin has sent several items of station news the high spot of which is that the Red Cross organisation is to run one of its periodic station tests on 26, 28 and 30 July at 0600 to 0700, 1130 to 1230, 1700 to 1800 and 2200 to 2300 on 7210. These test transmissions are to ensure that in the event of an emergency the Red Cross has a viable means of world-wide communication from Geneva. They will QSL if you report to 7 Ave. de la Paix 1211 Geneva 8. Robin also reports that the BBC is to supply Marconi SW transmitters and aerials to Masirah Island station by December and that Russia is supplying four 250kW transmitters to Radio Prague between 1977 and 1980.

Our newcomers take the floor now led by Jeffrey Raynor from Prestwich who runs a domestic radio/ cassette recorder. Despite the fact that this is hardly a communications set Jeffrey says that he has loads of fun listening to "the affairs of distant nations". He has connected a 25ft long wire and pulled in the following:---

Radio Pekin on 6100 at 0215 Radio Norway on 6050 at 0030 Radio Sweden on 6850 at 0030 Voice of Turkey on 9550 at 2300

I can assure Jeffrey that he has not as he suspects moved the cursor needle, the fact that his frequencies are not spot on are a result of the crude calibration of domestic receivers. This extra aerial, in my experience, can throw the calibration by affecting the oscillator. Lock on to a station of known frequency and then add the aerial. When this is done one finds the station again and notes the difference remembering to take this difference into consideration when logging in that band.

Marcus Duncan from Thornaby also notices this phenomena of calibration movement on his Alba and seventy-foot long wire. He has made a start on the PW General Coverage set with which project your column wishes him the best of luck. He logs as follows:—

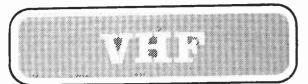
Israel Radio on 7412 at 2250 Voice of Iran on 9005 at 2015 Radio Cairo on 9700 at 2210 WYFR on 11645 at 1810

Two rarely reported stations come from our old friend Roy Patrick of Derby with his Trio 9R59 and Joystick. In this lean spell Roy cheers us all up by proving that some DX is still running by logging Emissor Regional Dos Acores (Azores) on 4865 and Radio Omdurman (Sudan) on 5039 at 1645. Coupled with this may I include another old friend Bob Burgess who writes but once a year but when he does he brings good cheer. From his Chelwood Gate OTH Bob and his Satellit 2000 bring a list 'as long as yer arm' of Latin Americans and Africans. While I admit that a Satellit is not a run-of-the-mill set it is still a portable and Bob uses it as a bedside set with all the hazards that local QRM can throw at him. I wish that I could list all of the loggings but space does not permit; however here is a handful:---

Radio Nacional Argentina on 11710 at 1900 Radio Nacional Brasilia on 11780 at 2300 Radio Nacional Chile on 9566 at 2300 Radio Rumbos Venezuela on 4970 at 2315 Conakry, Guinea on 4910 at 2245 Pointe Noire, Congo on 4843 at 2240

Bob mentions that Radio Nacional Chile is heavily jammed from Eastern Europe. Bob is hoping to retire shortly so we wish him well and wonder what sort of a logbook he will have, with more time to spend DXing.

We end this month on the subject of the Voice of America. J. H. Collick of Erdington would like to get in touch with VOA regarding their Big Band shows run by the veteran Willis Conover. The address to write to is VOA US Information Agency, Washington DC 20547 USA. This address also covers QSLs. With that I will wish you best 73s and all the best.



by Ron Ham

TE start 'down-under' with another interesting letter from Anthony Mann, Western Australia, who tells of some mouth-watering DX on Febuary 11 and 12, almost at the end of their sporadic-E "season". The 12th was most unusual, says Anthony, video signals on Ch.0 (46.25MHz) from eastern Australia (1900 miles) was received as early as 0800, local time. Later, between 1240 and 1400 video was received from New Zealand on 45.25MHz (3400 miles). Prior to this a strong signal from an Hawaiian amateur was heard on 10m (KH6AHL working VK8AC). Around 1800 (1000 GMT) the 10m band opened toward Europe, with both UK and French amateurs dominating the band. At 1037 GMT the German beacon DL0IGI (28.195MHz) was heard for less than a minute. Anthony uses a Barlow-Wadley XCR-30 with a folded dipole on 10m. While listening to an amateur broadcast he heard that, on January 24 and 25, the Great Australian Bight between Adelaide in VK5 and Albany in VK6 (1200 miles over water) had been bridged on 1296MHz. Our congratulations to the stations concerned.

In 14 years of routine observations the writer has never known a month like February when the VHF bands were so quiet. From the 1st to the 24th the atmospheric pressure did not rise above $30 \cdot 0''$, in fact for 10 of those days it was nearer $29 \cdot 5''$, so, at no time was there any 'life' in the troposphere for our radio signals. However, this has not daunted our readers. George Zitterstein, G8ITS, City of London, has now installed a barograph at his station, he was lucky to find one with an electric motor to drive the drum, to try and find out why the changing AP can become frequency critical within the VHF/UHF part of the spectrum.

Alan Baker, G8LGQ, our watchdog in Newhaven, noticed a 'lift' on 2m around 1500 on the 12th when he heard signals from PA0, but this did not last long and dashed the hopes of both Alan and Ernie Hoare, G8BDJ, Southwick, Sussex, of some good dx.

During the 500th edition of World Radio Club, broadcast by the BBC World Service on February 16th the team was asked. "When is sunspot minimum?" and this prompted your scribe to examine his records and put you all in the picture. After all, we are at the beginning of a new cycle and as it develops over the next few years, the increased solar activity will affect our mutual interests with DX again on the 10m band and more auroral activity to upset our VHFs. The following figures show the number of days each year that the writer's telescope recorded solar radio noise during the old sunspot cycle: - 1969-101, 1970-176, 1971-203, 1972-160, 1973-118, 1974-140, 1975-65, 1976-72. The peak activity was in 1971, followed by a downward trend until the end of 75 and the slight increase shown in 1976 may well be the start of the new cycle. Nigel Fisher, Co-ordinator of the Radio Astronomy Group of the South East Essex Astronomical Society, says that they have now completed their solar radio telescope and made satisfactory tests.

The solar radio waves will be collected on a Yagi aerial and fed to a Microwave Modules 136MHz converter; an Eddystone communications receiver is used as the IF amplifier and detector and an integrated circuit DC amplifier drives an Evershed and Vignoles pen recorder. Nigel's group will be observing between 1000 and 1400 hrs daily and will be supplying both the British Astronomical Association and ourselves with information as soon as they are properly organised.

Cmdr. Henry Hatfield, Sevenoaks, reports seeing a tiny solar flare and a plage through his spectrohelioscope on the 8th and on the 13th he saw a group of several sunspots. This accounts for the solar radio noise recorded by your scribe at 136MHz from the 10th to 13th inc. John Smith, Cranleigh, Surrey, noted that the mean solar noise level had increased at 140MHz over that period. John has been making radio observations of the sun for more than 12 years.

No doubt some form of solar activity was responsible for the ionospheric disturbance reported by the BBC World Service during the early hours of the 1st; this sort of information from the BBC is most useful to those readers who use the DX bands. Apart from a couple of short bursts of signals from the Cyprus beacon, 5B4CY, (28·180MHz) at 0830 on the 16th and 0905 on the 19th, the 10m band has been dead.

Did you know that the "hissing" which you can hear above the background noise of your receiver when the sun is "active", was first discovered in 1935 by Denis Heightman, G6DH, on the 10m band? Our readers are active in many fields. On February 18th Flt./Lt. John Keegan, CO 2464 (Storrington) Squadron, ATC, used 5 vehicles, each equipped with a Pye Cambridge, during a night exercise. These mobile units were spread over a 10 mile area of hills, vales, and woodland, not ideal for VHF communication with the short aerials attached to the cars. In addition the AP was low and falling and a moderate gale was blowing, but, despite this, and the undulating terrain, good communications were maintained throughout. On board each vehicle was a cadet wireless operator, and each one a keen radio enthusiast. After the exercise they were convinced that the "low-band" frequency was much better than the "high-band" frequency for reliable communications when conditions are bad.

The writer plans to be at the RSGB Convention at Alexandra Palace on May 7th and will be pleased to meet any of our readers who may be there. Thanks again to all who have sent in reports.

BROADCAST BANDS Short Wave Reports by the 15th of the n to Derek Bell, 169 Max Rd., Chaddeston, D Medium Wave Logs to Charles Molloy Segars Lane, Southport, PR8 3JG.)erby.
AMATEUR BANDS Logs covering any amateur band/s in alphabetical order by the 25th of the n to Eric Dowdeswell G4AR, Silver Leatherhead Road, Ashtead, Surrey, 2TW.	Firs,
VHF Reports on VHF matters to Ron Ham, Far Greyfriars, Storrington, Sussex RH20 4HE	aday,

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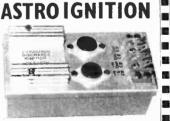
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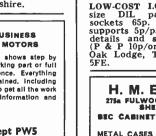
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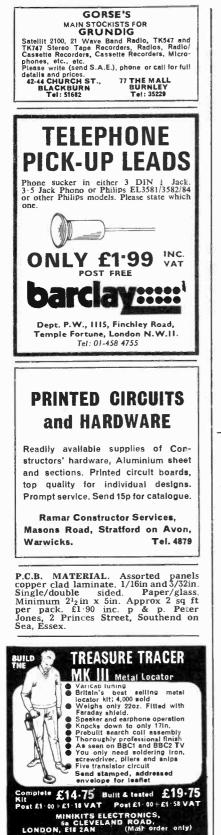
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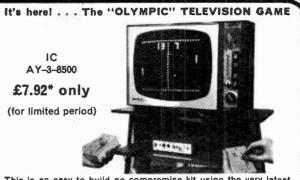
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POLYESTER CAPACITORS: Axial lead type. (Values are In µf). 400 V: 0-001, 0-0015, 0-0022, 0-0033 7p; 0-0047, 0-0068, 0-01, 0-015, 0-018 6p; 0-022, 0-033, 9p; 0-047, 0-068, 0-1, 11p; 0-15 0-22, 13p; 0-33, 0-47 28p; 0-68 38p. 160 V: 0-039, 0-15, 0-22 11p; 0-33, 0-47 19p; 0-68 1, 10 22p; 1-5 29p; 2-2 32p; 4-7 39p. DUBILIER: 1000 V: 0-01, 0-015 16p; 0-022 18p; 0-047 16p; 0-1 29p; 0-47 43p; 1-0, 68p.	AD161* 36 BC212L 13 BFR00 43 GC22 23 2N667* 21 2N3823* 49 AD162* 36 BC213 11 BFX18* 54 GC22 53 2N667* 21 2N3824* 89 AF114* 20 BC213L 13 BFX29* 28 GC33* 28 2N668* 39 2N386* 90 AF114* 20 BC214L 14 BFX38* 38 GC44* 44 2N506* 17 2N3903 18 AF115* 20 BC214L 16 BFX33* 40 GC44* 44 2N706* 17 2N3904 18 AF117* 23 BC328 18 BFX4* 38 GC139* 140 2N707* 50 2N3906 18 AF117* 27 BC328 18 BFX4* 38 GC139* 140 2N707* 50 2N3906 18	
POLYESTER RADIAL LEAD (Values In #), 259V: FEED THROUGH 0:01, 0:015 5p;0:022, 0:027 6p;0:033, 0:047, 0:068, 0:1 7p; 0:15 FEED THROUGH 10p;0:22, 0:33 12p;0:4714p;0:068 18p;1:021p;1:5 24p;2:2 28p, 1000pF/350V	AF121* 39 BC481- 40 BFX84* 24 CC141* 157 2N914* 18 2N4058 18 AF124* 30 BC462* 30 BFX85* 28 CC170* 40 2N916* 27 2N4051 13 AF125* 30 BC547 15 BFX85* 28 CC170* 40 2N916* 27 2N4051 13 AF104* 40 BC547 15 BFX85* 24 CC171* 44 2N41289 24	
ELECTROLYTIC CAPACITORS: Axial lead type (Values are in µF). 236V:100µF, 46p: 100V:20, 6p: 63V:10.47, 1.0, 1.5, 2.2, 2.5, 3.3, 4.7, 6.8 7p: 8, 10, 15, 22, 8p: 47, 32, 50, 10p; 63, 100, 12p; 50V:10, 10, 6p: 50, 100, 16p; 220, 10p; 470, 30p; 1000, 43p; 2200, 44p; 46V:22, 8p; 100, 10p; 3300, 52p; 35V:10, 33, 7p; 330, 470, 27p; 1000, 33p; 2200, 44p; 46V:22, 8p; 100, 10p; 2300, 52p; 35V:10, 33, 7p; 330, 470, 27p; 1000, 33p; 2200, 34p; 3300, 35p; 470, 04, 10p; 16V:10, 40, 47, 68, 8p; 100, 125, 7p; 470, 12p; 1000, 150, 15p; 2200, 25p; 16V:4, 100, 6p; 640, 10p; 1000, 14p, TAG-END TYPE: 76V: 4500 35p; 46V:2000 95p.	AF127* 30 BC348 15 BFX88* 28 OC202* 135 D1000* 18 D1487* 34 AF136* 30 BC546 15 BFX88* 28 OC202* 135 D1000* 18 D1487* 34 AF136* 70 BC537 28 BFY50* 17 OC203 150 D1961* 61 201613 12 AF178* 70 BC537 28 BFY51* 17 OC204* 150 211131* 19 2016136 12 AF136* 70 BCY32* 65 BFY52* 17 TIP29 43 211131* 23 2011318 12 AF181* 64 BCY32* 85 BFY52* 17 TIP29 43 21132* 23 201518 12 AF181* 64 BCY43* 85 P1792A 43 211342* 23 201518 12 AF183* 66 BCY40* 96	
TANTALUM BEAD CAPACITORS 38V: 0.1μF, 0.22, 0.33, 0.47, 0.68, 1.0, 2.2μF, 3.3, 4.7, 6.8, 25V: 1.5, 10. 20V POTENTIOMETERS (AB or EGEN) Carbon Track, 0.25W Log & 0.5W 15. 16V: 10μF, 47, 10V: 4.7, 15, 25, 33. 9V: 22, 47μF, 68, 3V: 100μF, 12p each. POTENTIOMETERS (AB or EGEN) Carbon Track, 0.25W Log & 0.5W	ASY20* 36 BCY70* 17 BSY95A 18 TIP31C 74 2N1307* 37 2N8458 36 BC107* 9 BCY72* 12 BL105* 168 TIP31* 52 2N1308* 48 2N8459 36 BC107* 9 BCY72* 15 E5567 48 TIP31* 54 2N1613 20 2N8458 42 BC107B* 10 BC115* 52 MD801* TIP31C* 58 2N1671* 150 2N6027 35 BC108* 12 BD12* 95 MJ400* 90 TIP32* 80 2N1671* 190 4031* 42 BC108E* 12 BD12* 95 MJ400* 90 TIP32* 80 2N16971* 90 4031* 114 BC108E* 12 BD12* 95 MJ400* 90 TIP32* 80 2N16971* 90 4031* 114	
MYLAR FILM CAPACITORS SLIDER POTENTIOMETERS 160V:0:001,0:002,0:005,0:014F 5p 0:015,0:02,0:03,0:04,0:05,0:0554F 6p 0:14F,0:15,0:27p. 50V:0:47µF 0:14F,0:15,0:27p. 50V:0:47µF 0:015,0:02,0:03,0:04,0:05,0:0554F 6p 0:14F,0:15,0:27p. 50V:0:47µF 0:015,0:02,0:03,0:04,0:05,0:0554F 6p 0:14F,0:15,0:27p. 50V:0:47µF 0:040,0:05,0:0504F 6p 0:040,0:0504F 6pm 0:0504F 6pm 0:0504F 6pm 0:0504F 6pm 0:0504F 6pm 0:0504F 6pm	BC109* 9 BD131* 38 MJ2855*120 TIP32E* 80 2N2217* 40 60317* 42 BC1096* 12 BD132* 38 MJ280*45 TIP32C* 32 2N218A*25 4032* 42 BC1096* 12 BD132* 38 MJ2370* 48 TIP32C* 32 2N218A*25 4032* 42 BC1096* 12 BD133* 60 MJ2370* 85 2N2219A*24 4032* 46 BC113 15 BD135 45 MJ257* 80 TIP33A* 95 2N2210A*24 4032* 46 BC113 15 BD135 45 MJ257* TIP33B*112 2N2221*21 4034* 43 BC115 18 BD137 42 MJE52* 74 TIP33C*120 2N222*2* 14034* 43 BC115 18 BD13* 47 MJE95***8* TIP34** 110 2N222*2* 14036* 36	
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Drive Unit complete motion motion drive 305p 305p HEAT DENCO C01LS with knobs and coupling 350p ⁻ 25; 50p 135; 20; 25; 50p 135; 20; 130p BINKS* Dual Purpose 'DP' Rng.1-5 B, Y, R, W 80p brit Ball Drive 100, 150p ⁻ 125; 70p	BC149 BF177* 28 MPSA7034 TIP3065*56 28/29260 8 Pair BC133 99 BF178* 28 MPSU0258 TIS43 27 2N29260 8 10p extra TTL 74+ 74155 76 4019AE 48 LINEAR IC*S MC1303L 148p 7400 14 7474 36 74157 95 4020AE 105 78L82AWC* 49p MC1303L 148p 7400 14 7474 36 74157 95 4021AE 105 706C 14 pln 710F 185p 7401 14 7475 37 7158 95 4021AE 105 700C 14 pln MC1310P 185p 7101 14 7475 37 74158 35 4021AE 105 700C 14 pln MC1310P 185p	
SILVER MICA (pF) CRYSTALS TO88 22p Rns. 1-5 B, Y, R. W 33. 4-7, 60, 68, 10, 12, 100 MHz 385p TO38 22p B9A Valve Base 23p 55, 100, 150, 150, 100 MHz 385p TO58 22p Rns. 1-5 B, Y, R. W 55, 100, 100, 150, 200 1MHz 385p TO58 22p RD2 32p 55, 100, 320, 380, 390, 1-5 1MHz 335p Insulation Kit RFC 7 (19mH) 90p 600 6 820 F 9p each 3 ³ / ₂ 2 ³ / ₂ 3 ³ / ₂ 10mH 3 ³ / ₂ S S	7402 18 7476 36 74160 116 4022AE 95 741C 8 pin 22p MC1458P* 77p 7403 16 7481 116 4023AE 18 747C 72p MC1458P* 77p 7404 20 7481 116 4023AE 18 747C 72p MC14509 91p 7404 20 7481 116 4023AE 75 743C 35p MC14009 91p 7405 22 7482 82 74163 116 4025AE 17 753 155p MFC6040* 97p 7405 44 7485 74165 120 4026AE 175 8038CC* 345p MK50253* 550p 7407 44 7484 95 74165 120 4026AE 175 8038CC* 345p MK50253* 550p 7407 44 7484 95 74165 120 4026AE 175 <td< td=""></td<>	
CERAMIC TRIMMER CAPACITORS 2-70F: 4-150F: 6-250F: 3-30pF 2-70F: 4-150F: 20p Capacitor	7400 22 7486 36 74167 382 4028AE 95 A'1-7-031 1229 NEBS0 7410 15 7489 390 7417 178 4030 AE 92 A'1-7-224 3489 NEBS0 410 15 160 17 174 173 103 AC3 12500 7724 175 174 173 17	
MINIATURE TYPE: S.A.E for list. EARINATIONES Sockets Screen type 3:5-60F; 3:00F; S.DEC 185p; Low Red. Black, Yellow 21p 5-25p; 6:00F; 88pF 30; U-DEC 'A' 375p; 3:5mm 18p [Texas] Yellow 21p COMPRESSION U-DEC 'B' 655p; 18p [Texas] 12p [ATTERY Yellow 21p	7416 35 7448 85 74177 116 4042AE 85 CA3018 137p NE865A= 175p 7417 33 7485 73 74180 106 4042AE 85 CA3018 72p NE865A= 175p 7420 16 7496 82 74181 299 4043AE 95 CA3020 145p NE867A= 176p 7421 26 7487 27 74180 289 4043AE 95 CA3023 145p NE867A= 182p 7421 27 7487 282 74181 289 4043AE 95 CA3023 145p NE867A= 182p 7422 24 74100 133 74184 184 4045AE 140 CA3028A* 195p ROM27102-2 250p 7423 36 74184 184 4045AE 1427 CA3035 145p ROM2513* 759p 7423 36 74185 1	
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TRANSFORMERS* (Mains Prim. 220-240V) PW PROJECTS 8-0-6V 100mA 90p 15-0-15V 1A 245p+ Easibuild Organ, General Coverage 9-0-9V 75mA 95p 8-0-18V 1A 275p+ Easibuild Trans. General Coverage	7437 30 74120 195 74196 118 4056AE 134 CA3081 1900 TAA651A 1930 7438 33 74121 34 476167 118 4056AE 112 CA3081 1900 TAA651A 1930 7440 17 74122 50 74198 248 4067AE 350 CA3090AQ 3950 TAA651A 1350 7441 7 74122 50 74198 248 4067AE 350 CA3090AQ 3950 TAA60A 1350	
12-0-12V 100mA 95p 30-0-30V 1A 295p + [Receiver, transistor lester, crimma- 15-0-15V 100mA 155p 6-0-8V 1-5A 345p + Chase, Stereo Touch Tuner. Tug-0- 0-12 0-12V 150mA 140p 6-0-8V 2A 270p+ War, Smoke Sensor Alarm, Reverb 0-6 0-6V 280mA 150 9-0-8V 2A 270p+ War, Smoke Sensor Alarm, Reverb	1442 14 143 150 140 150 140 150 <th 150<="" td="" th<=""></th>	
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OA9 9p OA47 8p OA70 8p OA79 12p OA81 10p OA85 12p	2A100V 44p 2A200V 46p 2A400V 53p 4A100V 72p 4A/200V 75p	3A200V 60 3A400V 110 3A600V 120 5A400V 120 7A400V 125	plass sided 5"x6" 64p	D BOARDS* dble/ sided S.R.B.P. 78p 7½**x8½** 160p 46p
OA30 6p OA31 6p OA31 6p OA20 8p OA200 8p OA202 8p IN914 4p	4A400V 79p 4A800V 129p 6A100V 73p 6A/200V 78p 6A400V 85p 6A/600V 99p	BT106 150p C106D 55p TIC44 25p TIC45 45p 2N4444 191p	FERRIC CHLO Anhydrous 85p DALO ETCH PEN* + spare t	+ 30p p. & p.
IN916 5p IN4001/2* 5p IN4003* 6p IN4004/5* 6p IN4006/7* 7p	BY164 56p ZENERS Rng:3·3V-33V 400mW 9p 1·3W 17p	TRIACS* 3A400V 113p 6A400V 114p 8A400V 160p 10A500V 195p 15A400V 200p	VEROBOARD* 0-1 (coppe 21 x 32" Mp 21 x 5" 43p	0-15 0-15 r clad) (plain) 28p 19p 39p 24p
1544 20p 3A/100V* 18p 3A/400V* 20p 3A/600V* 27p	VARICAPS MVAM2 135p BB104 40p BB105B 30p BB106 45p	15A500V 285p 30A400V 395p 40430 99p 40659 99p	3 x 3 " 43p 3 x 5 48p 2 x 17" 134p 3 x 17" 173p 4 x 17" 222p Pkt of 36 pins	40p — 53p 34p 107p 73p 143p 35p — 145p 210
3A/1000V*30p 6A/600V 50p	Noise Diode Z5J 105p	DIAC* ST2 25p	Spot face cutter Pin Insertion too	69m
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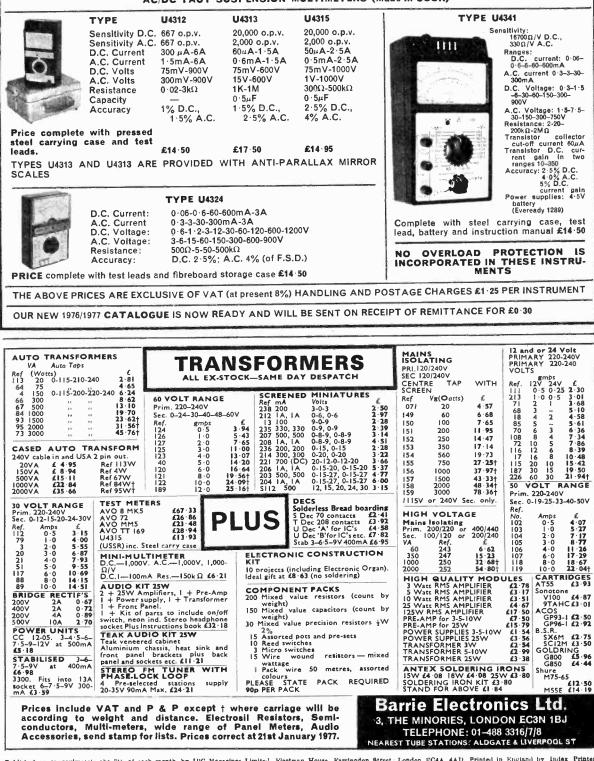
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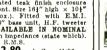
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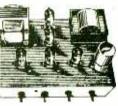


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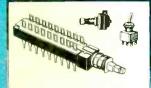
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