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| ADt49 | 68p | 1488 | 10p* | BD123 | 50 | ME4001 | 10p* | 4002 | p |
| AD181 | 52p | 1498 | $11 p^{*}{ }^{1 / 2}$ | 124 | 50p | 4101 | 11p* | 4005 | p |
| 162 | 52p | 154 | 14p* | 131 132 | 42p | MPF102 | 44p | 4148 |  |
| MCH/PR | 1.24 | 1578 | 12p* | . 138 | $54 \mathrm{p}{ }^{4}$ | - ${ }^{0}$ | 14 p | 2N2210 | 30p |
| AF116 | 24p | 177 | 17p | 155 | $7{ }^{\text {P }}$ * | 91 | 8 | 2846 | 45p |
| 117 | 23 p | 1788 | 18p | 8F158 | 20\% ${ }^{\text {\% }}$ | 200 | 10 p | 2926 | 13p* |
| 124 | 30p | 1798 | 19p | 184 | 25p | OC28 | 4.20 | 2928 G | 15p* |
| 239 | 48 p | 1848 | 12p* | 185 | $23 p$ | 29 | 1.20 | 3033 | ${ }^{25}{ }^{\text {p }}$ |
| AU113 | 1.72* | 1841 | 11p* | 194 | 40p* | 35 | 4.20 | 3055 | 00p |
| BC107 | 11p | 187 | 28p | 196 | 12p ${ }^{\circ}$ | 44 | 45p | 3702 | $11 p^{*}$ |
| 107A | 12p | 212 A | 43p* | 197 | 12p ${ }^{2}$ | 71 | $35 p$ | 3703 | 14p |
| 108 | 10p | 212 L | 15p: | 199 | 15p ${ }^{\circ}$ | ORP12 | ${ }^{88}$ | 3704 | 43 p |
| BC1088 | 11p | 2138 | 12p* | 200 | $3{ }^{3}$ | TIP29A | 470* | 3705 | 42p |
| 108 C | 12p | 214 | $15 p^{*}$ | $338$ | $34 p$ | 30 A | $\begin{aligned} & 55^{\circ} \\ & 57 p \end{aligned}$ | $3707$ $3819 \mathrm{E}$ | $\begin{aligned} & 12 p^{*} \\ & 25 p^{\circ} \end{aligned}$ |
| DIODES $50 V 3 A$ | 13p | $\begin{aligned} & \text { BRIDe } \\ & \text { 100V1A } \end{aligned}$ | 23p | $\begin{aligned} & \text { RESIS } \\ & \text { L WAT } \end{aligned}$ | $\begin{aligned} & \text { STORS } \\ & \text { TTE12 } \end{aligned}$ | $\begin{aligned} & 1 \% p^{*} \text { each } \\ & (5 \%) \end{aligned}$ | LINE | EAR I/C' | ${ }^{31}{ }^{24}$ |
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| ECC83 | 0.38 | GZ30 | 0.65 | PL500 | 0.85 | ${ }^{68887}$ | 0.55 3.00 | 30P12 | 1.00 |
| ECC84 | 0.35 0.45 | GZ32 | 0.65 0.75 | PL504 | 0.88 0.90 | 68S7 | 1.40 | $30 \mathrm{P1} 19$ | 0.95 |
| ECC83 | 0.50 | N309 | 1.50 | PL509 | 1.60 | 68W6 | 1.00 | 30PL1 | 0.95 |
| ECH35 | 1.50 | KT61 | 3.40 | PL802 | 2.00 | W7 | 1.00 | 30PL13 | 1.10 |
| ECH42 | $0 \cdot 85$ | KT66 | $3 \cdot 40$ | PX25 | $3 \cdot 50$ | 6 C 4 | 0.40 | 30PL14 | 1.10 |
| ECH81 | $0 \cdot 35$ | K T81 |  | PY33 | 0.63 | ${ }_{6} \mathrm{CO} 6$ | $1-80$ | 35 W 4 | 70 |
| ECH83 | $0 \cdot 50$ | （7C5） | 2.00 | PY81 | － 50 | ${ }^{6} \mathrm{CH} 6$ | 2.20 | 35Z4GT |  |
| ECL80 | $0 \cdot 60$ | KT88 | 4.00 | PY82 | 0.45 | 6CW4 | 00 | 50CO6G | $1 / 20$ |
| ECL89 | 0.42 | KTW61 | 1.50 | PY83 | 0.50 | 6F23 | 0.90 | 807 | 00 |
| ECL83 | 0.75 | MU14 | 1.00 | PY88 | 0.63 | 6 625 | 1.00 | 8131 TT | 17．50 |
| ECL86 | 0.55 | N78 | 7.50 | PY500 | 1.18 | 6 F 28 | 0.75 | 813US |  |
| EF37A | 1.50 | OA2 | 0.45 | PY81／800 | 0.50 | 6.51 M | 065 | 66 A |  |

EXPRESS POSTAGE

15p per arder In UK

AA119 0．07／BD131 0．

| AA119 | 0.07 | BD131 | 0.40 |
| :--- | :--- | :--- | :--- |


| AAZ15 | 0.18 | BD132 | 0.4 |
| :--- | :--- | :--- | :--- |


| AC107 | 0.72 | BF115 | 0.20 |
| :--- | :--- | :--- | :--- |
| AC128 | 0.25 | BF173 | 0.2 |
| AC1 | 0.2 |  |  |

$\begin{array}{llll}\text { AC128 } & 0.25 & \text { BF173 } & 0.2 \\ \text { AC127 } & 0.25 & \text { BF179 } & 0.30\end{array}$

| AC128 | 0.22 | BF179 | 0.30 |
| :--- | :--- | :--- | :--- |
| AC176 | 0.27 | BF181 | 0.30 |
| AC18 | 0.35 | BF194 | 0.10 |

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\begin{array}{ll|ll}
\text { BCD11 } & 1.55 & \text { OA81 } & 0.15 \\
\text { BD124 } & 0.75 & \text { OA91 } & 0.07 \\
\hline
\end{array}
$$

| XN7400 | $0 \cdot 16$ | SN7428 | 0.40 | SN7486 | 0.47 | SN74145 | 1－26 | XN74192 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SN7401 | 0.16 | SN7430 | 0.16 | SN7490 | 0.55 | SN74150 | 1.75 | SN74193 | $2 \cdot 00$ |
| SN7402 | 0.16 | SN7433 | 0.37 | SN7491 AN |  | SN74151 | 1.00 | SN74194 | $1 \cdot 30$ |
| SN7403 | 0.16 | SN7437 | 0.37 |  | 1.00 | SN74154 | 2.00 | SN74195 | 1.10 |
| SN7404 | $0 \cdot 25$ | SN7438 | 0.37 | SN7492 | 0.70 | SN74155 | 1．00 | SN74196 | 1． 20 |
| SN7405 | 0.22 | SN7440 | $0 \cdot 22$ | SN7493 | 0.70 | SN74156 | 1.00 | SN74197 | $1 \cdot 20$ |
| SN7406 | 0.42 | SN7441A |  | SN7494 | 0.80 | SN74157 | 0.85 | XN74198 | $2 \cdot 7$ |
| SN7407 | 0.42 |  | 0.92 | SN7495 | 0.80 | SN74170 | 2.52 | SN74199 |  |
| SN7408 | 0.28 | SN7442 | 0.79 | SN7496 | 0.95 | SN74174 | 1.57 |  |  |
| SN7409 | 0.28 | SN7450 | 0.16 | SN7497 | 3.87 | SN74175 | 1.10 |  |  |
| SN7410 | 0.16 | SN7451 | 0.16 | SN74100 | 1.89 | SN74176 | 1.26 |  |  |
| SN7411 | 0.25 | SN7453 | $0 \cdot 16$ | SN74107 | 0.45 | SN74190 | 2.00 |  |  |
| SN7412 | 0.30 | SN7454 | 0.16 | SN7110 | －58 | SN74191 |  |  |  |
| SN7413 | 0.36 | SN7460 | 0.16 | SN74118 | $0 \cdot 90$ |  |  |  |  |
| SN7416 | 0.36 | SN7470 | 0.36 | SN74119 | 1.68 | DIL |  | 8 pin | $5 p$ |
| SN7417 | 0.38 | SN7473 | 0.41 | SN74121 | 0．50 | D |  |  |  |
| SN7420 | 0.16 | SN7474 | 0．42 | SN74122 | $\begin{aligned} & 70 \\ & 1.00 \end{aligned}$ | － |  |  |  |
| SN7422 SN7423 | 0.25 0.37 | SN7480 SN7483 | 0.60 1.10 | SN74123 SN74141 | 1.00 0.90 | SOCK | ET | 16 p | $7 p$ |

## VAT

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| OA200 | 0.10 | ZTX501 | 0.13 | 2N2005A |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| o azor | 0.10 | ZTX503 | $0 \cdot 1$ | 2N2908 | 0.22 |
| ${ }^{\circ} \mathrm{C} 16$ | $1 \cdot 25$ | ZTX531 | － 25 | 2N2928 | － 15 |
| OC20 | 2.60 | ZTX550 | $0 \cdot 10$ | 2N3033 | － 20 |
| OC23 | 2.25 | IN014 | 0.06 | 2N3055 | $0 \cdot 6$ |
| OC25 | 0.85 | 1N4001 | 0.07 | 2N3525 | － 90 |
| OC28 | 0.75 | 1 N4002 | 0.07 | 2N3614 | 00 |
| OC35 | 0.75 | 1 N 4003 | 0.08 | 2N3615 | 1.50 |
| 0 C 36 | － 75 | 1 N 4004 | 0.08 | 2N3702 | $0 \cdot 13$ |
| OC42 | 0.50 | 1 N 4005 | 0.09 | 2N3703 | －13 |
| OC44 | 0.45 | 1 N4006 | 0.19 | 2N3704 | －15 |
| OC45 | 0.45 | 1N4007 | － 11 | 2N3705 | － 13 |
| OC71 | 0.25 | 1N4009 | － 08 | 2N3706 | － 13 |
| OC72 | 0.45 | 1N4148 | $0 \cdot 08$ | 2N3707 | $0 \cdot 13$ |
| OC76 | 0.45 | IS921 | － 0 ¢ | 2N3708 | $0 \cdot 6$ |
| OC77 | 0.75 | 1S2033 | 0．20 | 2N3709 | － 12 |
| OC81 | 0.50 | IS2051A | － 20 | 2N3710 | － 13 |
| OC81D | 0.28 | IS2100A | 0． 20 | 2N3711 | － 13 |
| OC812 | 1.00 | IS3010 | $0 \cdot 25$ | 2N3819 | － 37 |
| $0 \mathrm{CB3}$ | 0.55 | 2N696 | － 20 | 2N3820 | － 55 |
| OC140 | 1.50 | 2N697 | －$\cdot 17$ | 2N3823 | － 0 |
| OC170 | 0． 50 | 2N706 | － 12 | 2N3903 | － 15 |
| OCT71 | $0 \cdot 50$ | 2N706A | 0.12 | 2N3904 | － 15 |
| OC200 | 1.00 | 2N1131 | 0.23 | 2N3905 | － 17 |
| OC201 | 175 | 2N1132 | 0． 25 | 2N3906 | － 17 |
| OC202 | 1.50 | 2N1302 | － 30 | 2N4058 | － 16 |
| OC203 | 1.50 | 2N1303 | － 40 | 2N4059 | － 12 |
| OCP71 | 1.25 | 2N1304 | 0.45 | 2N4080 | $0 \cdot 13$ |
| ORP12 | 0.60 | 2N1305 | 0.45 | 2N4081 | 0.13 |
| ORP60 | － 65 | 2N1306 | 0.50 | 2N4062 | － 14 |
| Tic44 | 0.30 | 2N1307 | 0.50 | 2N4289 | － 30 |
| TIC226D | 1.40 | 2N1308 | 0.50 | 3N125 | $1 \cdot 75$ |
| TIL209 | 0.22 | 2N1309 | 0.50 | 3N141 |  |
| TIP42A | 0.70 | 2N1613 | 0.21 | 40360 | 0.40 |
| ZTX107 | $0 \cdot 10$ | 2N2147 | 1.20 | 40361 | － 45 |
| ZTX108 | 0.10 | 2N2160 | 0.75 | 40362 | $0 \cdot 40$ |
| ZTX300 | 0.12 | 2N2369 A | － 25 | 40430 | － |
| ZTX301 | 0.13 | 2N2646 | 0.50 |  |  |
| ZTX302 | 0.17 | 2N2904 | － 25 |  |  |
| ZTX304 | 0.22 | 2N2904A | 0.25 |  |  |
| ZTX500 | 0.12 | 2N2905 | 0.30 |  |  |
| SN7486 | 0.47 | SN74145 | 1－26 | XN74192 | $2 \cdot 00$ |
| SN7490 | $0 \cdot 55$ | SN74150 | 1.75 | SN74193 | 2.00 |
| SN7491 A |  | SN74151 | 1.00 | SN74194 | $1 \cdot 30$ |
|  | 1.00 | SN74154 | 2.00 | SN74195 | 1.10 |
| SN7492 | 0.70 | SN74155 | 1.00 | SN74196 | $1 \cdot 20$ |
| SN7493 | 0．70 | SN74156 | 1.00 | SN74197 | 1.20 |
| SN7494 | 0.80 | SN74157 | 0． 95 | XN74198 | $2 \cdot 7$ |
| SN7495 | 0.80 | SN74170 | 2.52 | SN74199 | $2 \cdot 52$ |
| SN7496 | 0.95 | SN74174 | 1.57 |  |  |
| SN7497 | 3.87 | SN74175 | 1.10 |  |  |
| SN74100 | 1.89 | SN74176 | $1 \cdot 26$ |  |  |
| SN74107 | 0.45 | SN74190 | 2.00 |  |  |
| SN7410 | － 58 | SN74191 | 2.00 |  |  |
| SN7418 | $0 \cdot 90$ |  |  |  |  |
| SN7419 | 1.68 | DIL |  | 8 pin | 15p |
| SN74121 SN74122 | 0.50 0.70 | OIL |  | 14 pin | 15p |
| SN74123 | 1.00 | SOC | E | 16 pi | $7 p$ |
| SN74141 | 0.90 | So | E | 16 pi | P |

## THIS MONTH＇S



SPECIAL OFFER

## SEMICONDUCTORS-COMPONENTS

| Type | Price | Type | Price |  | Price |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AA129 | ¢0.08 | BY100 | co. 18 | BYZ11 | E0. 31 | OA91 | ¢0. 07 |
| AAY30 | £0.09 | BY107 | 20.12 | BYZ12 | £0.31 | OA95 | 50.07 |
| AAZ13 | £0.10 | BY105 | E0. 18 | BYZ13 | E0. 26 | OA182 | £0. 07 |
| AAZ17 | £0.10 | BY114 | c0. 12 | BYZ16 | £0.41 | OA200 | 50.08 |
| BA100 | co. 10 | BY124 | - 50.12 | BYZ17 | $\mathrm{EPO}_{50} 36$ | OA202 | co. ot |
| BA102 | ¢0. 32 | BY126 | - 50.15 | BY218 | £0.36 | SD10 | c0. 06 |
| BA148 | £0.15 | BY127 | - 20.16 | BYZ19 | £0. 28 | SD19 | c0. 06 |
| BA154 | £0.12 | BY128 | - 50.16 | OA10 | £0.35 | IN34 | c0. 07 |
| BA155 | E0. 14 | BY130 | -E0. 17 | OA47 | E0.07 | IN34A | c0. 07 |
| BA156 | ¢0. 14 | BY133 | - 50.21 | OA70 | ¢0.07 | IN914 | c0. 06 |
| BA173 | £0.15 | 8 BY 164 | £. 51 | OA79 | £0.07 | IN916 | c0. 06 |
| B8104 | E0 15 | BY176 | - 50.75 | OA81 | £0.07 | IN4148 | c0. 06 |
| BAX13 | £0.07 | BY206 | £0.00 | $0 \mathrm{OA}^{85}$ | £0.09 | IS44 | £0.05 |
| BAX16 | co 08 | BYZ10 | ¢0. 36 | OA90 | £0.07 | IS9 | c0. 08 |
| SILICON RECTIFIERS |  |  |  |  |  |  |  |
| Type | $\begin{aligned} & \text { Price } \\ & 80.06 \end{aligned}$ | Type IN4003 | $\begin{aligned} & \text { Price } \\ & £ 0.07 \end{aligned}$ | $\begin{aligned} & \text { Type } \\ & \text { IS020 } \end{aligned}$ | $\begin{aligned} & \text { Price } \\ & £ 0.10 \end{aligned}$ | Type IS031 | $\begin{aligned} & \text { Price } \\ & \mathrm{E} 0 \cdot 25 \end{aligned}$ |
| IS921 | ¢0.07 | IN4004 | £0.08 | IS021 | E0.11 | in5400 | c0. 13 |
| IS822 | $\pm 0.08$ | in4005 | £0. 09 | IS023 | c0. 13 | IN5401 | £0.15 |
| IS923 | E0.09 | IN4006 | c0. 10 | IS025 | co. 14 | IN5402 | £ 0.16 |
| IS924 | £0.90 | IN4007 | £0. 11 | IS027 | co. 18 | IN5404 | co. 17 |
| IN4009 | E0.05 | IS015 | c0. 09 | IS029 | c0. 20 | IN5406 | c0. 21 |
| IN4002 | 20.06 |  |  |  |  | IN5407 | ¢0.25 |


| 2 AMP TOS CASE |  |  | 10 AMP TO4\% CASE |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Volts | No. | Price | Volts | No. | Price |
| 100 | TR12A/100 | EOP 31 | 100 | TR110A/100 | £0.77 |
| 200 | TRT2A/200 | 60. 51 | 200 | TR1 toA/200 | ¢0. 92 |
| 400 | TR12A/400 | ¢0. 71 | 400 | TR110A/400 | ¢1.12 |
| - AMP TOES CASE |  |  | IO AMP TO4\% CASE |  |  |
| Volts | No. | Price | Volts | No. | Price |
| 100 | TR16A/100 | co. 51 | 100 | TR110A/100 | E0-77 |
| 200 | TR16A/200 | 60. 61 | 200 | TR110A/200 | E0. 92 |
| 400 | TR16A/400 | C0. 77 | 400 | TR110A/400 | ¢1. 12 |

OIE CASE

| 600mA TOIE CASE |  |  |
| :---: | :---: | :---: |
| Volts | No. | Price |
| 10 | THY600/10 | c0. 13 |
| 20 | THY600/20 | ${ }_{20} 0.13$ |
| 30 | THY600/30 | ¢0.19 |
| 50 | THY600/50 | co. 22 |
| 100 | THY600/100 | 80.25 |
| 200 | THY600/200 | ¢0. 38 |
| 400 | THY600/400 | ¢0. 45 |
| 1 AMP TOS CASE |  |  |
| Volts |  | Price |
| 50 | THY1A/50 | £0. 26 |
| 100 | THY1A/100 | $\pm 0.27$ |
| 200 | THY1A/200 | E0.28 |
| 400 | THY1A/400 | c.0. 36 |
| 600 | THY1A/600 | £0.45 |
| 800 | THY1A/800 | E0.58 |


| 3 AMP TO66 CASE |  |  | $\begin{aligned} & \text { Volts } \\ & 50 \\ & 100 \\ & 200 \end{aligned}$ | No. THY16A/50 | Price <br> C. 0.54 |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |
| Volts | No | Price |  | THY16A/200 | E0. 62 |
| 50100 | THY3A/50 | c0. 25 | 400 | THY16A/400 | c0. 77 |
|  |  | c0. 27 | 600 | THY16A1600 | CO. 90 |
| ${ }_{20}^{10}$ | THY3A/200 | 20. 33 | 800 | THY16A/800 | £1.39 |
|  | THY3A/400 <br> THY3A/600 <br> THY3A/800 | $\begin{aligned} & £ 0.42 \\ & £ 0.50 \\ & 80.65 \end{aligned}$ |  | AMP TO94 CASE |  |
|  |  |  | 30 |  |  |  |
|  |  |  | ${ }^{\text {Voits }}$ | No. <br> THY30A/50 | Prl |
|  |  |  |  |  | c1. 18 |
| 5 AMP TO6s CASE |  |  | ${ }^{2} 200$ | THY30A/200 |  |
| Volts | No. | Price | 400 | THY30A/400 | E1.79 |
|  |  | ¢0. 36 | 600 | THY30A/600 | E.3.50 |
| 50 | THY5A/50 | £0.48 |  |  |  |
| 200 | THY5A/200 | ¢0. 50 |  |  |  |
| 400 600 | THY5A/400 | ¢0.57 | No. ${ }^{\text {BTios/500R }}$ |  | c0:80 |
| 600800 | THY5A/800 |  | ${ }_{\text {BT106 }}$ BTi02/50R |  | coso |
|  |  |  |  |  | E1. 25 |
|  |  |  | BT108 |  |  |
|  |  |  | 2 N 32282 N 3525 |  | ci. 98 |
| 5 AMP TO220 CASE |  |  |  |  | ¢0.77 |
| 400600 | THY5A/400P |  | 2N3525 |  | E0. 33 |
|  | $\begin{array}{ll} \text { THY5A/600P } & \text { ¢0. } 69 \\ \text { THY5A/B00P } & \text { CO. } \end{array}$ |  | BTX30 | 100 | E0. 46 |
| 800 |  |  | C106/4 |  | E0.60 |

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Bassorted types. SL403, 76013, 76003, etc.
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VBt-Approx.ORDER No. 16199 60pB2-Approx. 30sq.in varlous sizes. 0.15 inORDER No. 16200

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C2-Values from $10 \mu \mathrm{~F}$ to $100 \mu \mathrm{~F}$
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EC3-Values from $100 \mu \mathrm{~F}$ to $680 \mu \mathrm{~F}$. 00 p

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R3- 60 mixed $i W 10-82 \mathrm{ka}$
R3-60 mixed iW $10-82 \mathrm{ka}$.
ORDER No. 16215
R4-60 mlxed it $W$ 100-8205 $\Omega$.
R5-40 mixed $+W$ W 100-820n.
ORDER No. 16217
R6-40 mixed iW 1-8.2kD.
ORDER No. 16218
RT 40 mixed $W 10-82 \mathrm{k} \Omega$.
R7- mixed $1210-82 \mathrm{k} \Omega$.
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R8-40 mixed it $100-820 \mathrm{kD}$.
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R9-60 mixed iW 1-1
ORDER No. 16230
R10-40 mixed $+W$ 1-10MD.
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| :--- | :--- | :--- |
| $0-50 \mu \mathrm{~A}$ | 1307 | $£ 3.50$ |
| $0-100 \mu \mathrm{~A}$ | 1308 | $£ 3.50$ |
| $0-500 \mu \mathrm{~A}$ | 1309 | $£ 3.50$ |
| $0-1 \mathrm{~mA}$ | 1310 | $£ 3.50$ |
| $0-50 \mathrm{~V}$ | 1311 | $£ 3.50$ | <br> MR2P TYPE Size $42 \times 42 \times 30 \mathrm{~mm}$ <br> | Value | No. | Price |
| :--- | :--- | :--- |
| O-50 | 1313 | $£ 4.80$ |
| $0-1 \mathrm{~mA}$ | 1315 | $\mathbf{£ 3 . 2 0}$ | <br> EDGEWISE <br> Size $3 \frac{1}{2} \times 1 \frac{1}{} \times 2 \mathrm{tin}$ Cut out $2 \geq 1 \times 1$ in Value No. Price $\begin{array}{lll}{ }_{0}^{0-1 m A} \\ 0-500 \mu A & 1317 & \text { E4.05 }\end{array}$ <br> MINIATURE BALANCE/TUNING METER <br> Size $23 \times 22 \times 26 \mathrm{~mm}$ Sensitivity $100 / 0 / 100 \mathrm{~mA}$ <br> | No. | Price |
| :--- | :---: |
| 1318 | £ 1.95 | <br> BALANCE/TUNING Size $45 \times 22 \times 34 \mathrm{~mm}$ Sensitivity $100 / 0 / 100 \mu \mathrm{~A}$ <br> | No. 1319 | $\begin{aligned} & \text { Price } \\ & \text { £2.00 } \end{aligned}$ |
| :---: | :---: |
| MIN. LEVEL METER |  |
| Size $23 \times 22 \times 26 \mathrm{~mm}$ |  |
| Sensitivity |  |
| No. 1320 | Price <br> £1.95 |
| Vu METER |  |
| Size $40 \times 40 \times 29 \mathrm{~mm}$ <br> Sensitivity $130 \mu \mathrm{~A}$ |  |
| No. $1321$ | Price <br> £2.00 |
| MM |  |

Size $60 \times 24 \times 90 \mathrm{~mm}$
Sensitivity 1000 ohms/V
A.C. VOLTS $0-10,50,250$ 1000
D.C. VOLTS $0-10,50,250$ 1000
D.C. CURRENT $0-1100 \mathrm{~mA}$

Resistance 0-150ks) | No. | Price |
| :--- | ---: |
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| Type Price | Type Price |
| :---: | :---: |
| ${ }_{\text {AC127 }} \mathbf{£ 0 . 1 4}$ | BC147 E E0.09 |
| AC128 $\quad$ £0.12 | BC148 - 50.09 |
| AC128K £. 0.26 | BC149 * 00.09 |
| AC132 $£ 0.15$ | BC157 *-E0-12 |
| AC134 £0-15 | BC158 * 0 - 12 |
| AC137 £0.15 | BC159 - E0-12 |
| AC141 $£ 0 \cdot 18$ | BC167 ${ }^{\text {ce }} 12$ |
| AC141K £ $0 \cdot 30$ | BC168 * 8.12 |
| AC142 £0-18 | BC169 * E0-12 |
| AC176 | BC169C |
| AC176K £0. 26 | C0. 12 |
|  | BC170 ${ }^{\circ} \mathrm{EO} 10$ |
| AC179 £0. 25 | BC171 - £0. 10 |
| AC180 £0. 20 | BC172 ${ }^{\circ} \mathrm{E0} 0 \cdot 10$ |
| А С180K 50.30 | BC173 *0.12 |
| AC181 E. 20 | BC177 £0.16 |
| AC181K ¢0. 30 | BC178 £0.16 |
| AC187 ¢0.16 | BC179 £0.16 |
| AC187K ¢0. 26 | BC180 £0.25 |
| AC188 ¢0.16 | BC18t ${ }^{\text {E0. }} \mathbf{2 5}$ |
| AC188K £0. 26 | BC182L |
| AD140 £0.60 | - $20 \cdot 10$ |
| AD142 \$0.85 | BC183 * $£ 0 \cdot 10$ |
| AD143 £0.75 | BC183L |
| AD149 £0.60 | - ¢0-10 |
| AD161 $£ 50 \cdot 36$ | BC184 * $£ 0 \cdot 10$ |
| AD162 $\quad$ £0. 36 | BC184L |
| AD161/162MP | ${ }^{+2} \mathrm{cos} 10$ |
| $£ 0.75$ | BC207 - E0-11 |
| AF114 E0.20 | BC208 ${ }^{-10.11}$ |
| AF115 £0. 20 | BC209 - 50.12 |
| AF116 $\mathbf{x} \mathbf{0} \mathbf{0} \mathbf{2 0}$ | BC212 'E0-11 |
| AF117 50.20 | BC212L |
| AF118 $£ 0.40$ | ¢0.11 |
| ${ }_{\text {AF124 }} \times 0.30$ | BC213 ${ }^{\mathbf{E} \mathbf{8 0 . 1 1}}$ |
| AF125 $\quad \mathbf{8} 0.30$ | BC213L |
| AF126 EO 0 | - £0.11 |
| AF127 E0. 32 | BC214 * ¢0.12 |
| AF139 <br> 10.58 | BC214L |
| AF180 C0.58 | ${ }^{*}$ E0. 12 |
| AF181 £0.5 | BC237 $£ 0.16$ |
| AF186 C0.58 | BC238 ¢0.16 |
| AF239 £0.38 | BC251 * ¢0. 15 |
| AL102 E0.95 | BC251A |
| AL103 £0.95 | *E0-16 |
| AU104 £1.00 | BC301 £0.30 |
| AU110 £1.00 | BC302 £0.28 |
| AU113 $\quad$ £1-00 | BC303 £0.32 |
| BC107A $£ 0.08$ | $8 \mathrm{C3O4}$ £0.34 |
| BC107B £0.08 | BC327 - E0. 16 |
| BC107C £0.05 | BC328 - E0. 15 |
| BC108A $£ 0.08$ | BC337 - 20.15 |
| BC108B 20.08 | ВС338 - ¢0.15 |
| BC108C E0.08 | BC440 $\quad 60.30$ |
| BC109B $£ 0.08$ | BC441 E0.30 |


$\begin{array}{lr}\text { Type } & \text { Price } \\ \text { Tip29C } \\ \text { £ } 0.62\end{array}$

## 74 SERIES TTL ICs



AMONG our constructional projects this month is a Gas/ Smoke Sensor Alarm that is designed to give a visual and audio warning when the gas or smoke content of the atmosphere rises above a predetermined level. In view of the apparent spate of explosions around the country in recent times, frequently due to the leakage of natural gas, it might be thought that we are trying to 'cash-in' on other people's misfortunes.

Nothing could be further from the truth. The project was scheduled for publication in this issue several months ago. Observant readers may have noticed that issues with major projects in them are being interspersed with issues containing a number of smaller constructional designs. The Gas/Smoke Alarm is just one of those designs, among a number that we shall be presenting to our readers in the forthcoming months.
Should any of our readers make up this Alarm gadget then we sincerely hope that it never goes off! It's like an insurance policy, really. One pays and pays but hopes that the need to collect will never arise! On the other hand, the Dual Power Supply also presented in this issue is something that ought to be seen to be working all the time.
There is nothing more annoying than to finish an interesting and enjoyable project only to find that there are not enough batteries in the shack to get your pride and joy working. Our design comprises two completely independent supplies, each fully monitored for voltage and current output, which should prove to be just about the most useful bit of gear to have around.
On a slightly different tack, the generally forecast shortage of components, and the inevitable increase in prices for those that are available, should tend to favour the smaller electronic project in the future. Every constructor, if he is wise, soon begins to collect bits and pieces, in the way of capacitors, resistors etc and every nut and washer should be preserved. There are those among us who can remember stripping down government surplus radio and radar equipment after the last war and getting jars full of lovely cadmium plated bolts, washers, nuts and so on, plus many hundreds of various components, not to mention untold numbers of that elegant and ubiquitous valve, the EF50! They must have been produced by the million!
'Ah!' I hear you saying 'but we can't get that sort of stuff today'. No, maybe not, but is there a home, house, shed or attic in the land that does not have a discarded TV in some dark corner? But whatever you do, be careful when handling the tube, especially when removing it from the chassis. If tube is sound and you intend to use it again, all well and good, but if you are going to dispose of it to the dustman then place it face downwards in a substantial cardboard box and mark it accordingly. DO NOT try to break it yourself or the resulting implosion could cause injury to the eyes or face.
From then on it is an enjoyable few hours work, pulling the set to pieces and sorting the 'swag' into types of components as you go along. Later, every component ought to be checked for actual value before being used again. This is where a lack of test equipment begins to show up. One of the more instructive and rewarding of pastimes is the building and using of test gear because when there is something wrong with a component or circuit it needs a little brain work to find the fault and that is where the learning begins.

ERIC DOWDESWELL Assistant Editor

## SILENT KEY

Readers will be sorry to learn of the death of Austin Forsyth, OBE, G6FO, in his 70th year. He had been Editor of Short Wave Magazine since 1938

## Fair enough

FOLLOWING on from the story in the February issue concerning the cancellation of the 1976 Audio Fair, we are now pleased to see the return of this major event in 1977. Due to take place at Olympia from September 12th to the 18th, this latest show promises to be an even greater success than those in past years.
The organisers, Iliffe Promotions Ltd., say that "although the show will retain its appeal to the specialist audio and $\mathrm{Hi}-\mathrm{Fi}$ enthusiasts, the 1977 Audio Fair will take account of the greatly increased family demand for sound and visual home entertainment".
Provisional bookings have already been taken up by such firms as BSR, Shure, National Panasonic, Eagle, Hitachi and JVC. It would appear that this is just the beginning of the list of exhibitors and that 'on the day' all the old firms, together with many new ones, will again be under the one Olympia roof.
Audio Fair, Iliffe Promotions Ltd., Dorset House, Stamford Street, London SE1 9LU.

## No free plugs for

 this productWHY is it that we British always seem to knuckle under to ideas and technical innovations, when all and sundry KNOW that our standards in the field of electrical safety are second to none? The point that we're getting at, is the recent revelation from those Barons of the EEC, that there will shortly be a new "World Plug", designed specifically to be compatible in any corner of the world. Fine ideas indeed you may say, but no amount of persuasion will convince these EEC commissioners and their bevy of technical advisers to incorporate a fuse within the plug. You see the plug is a 16 A moulded de-
vice with three in-line flat pins, permanently fixed to the lead and totally fuseless. This of course would mean that all equipment purchased would have to be fused in the unit itself. and

should you decide that you would like to keep your "antiquated" 13A system, recently installed at the expense of next year's holidays, you will be delighted to know that any new equipment would have to have its plug physically cut off and a newsorry old-plug put on.

Another controversial, and in our opinion downright dangerous idea, is that generally, the sockets for these plugs will be without shutters, although the commissioners do state that "provision MAY be made in the UK for the use of shuttered socket outlets"-how thoughtful of them! Maybe if they do, we won't have a spate of dead children-the result of poking knitting needles and the like into the open holes. It's a pity we can't say the same thing for those unsuspecting French or German children. Still we suppose that the Commissioners have thought all about this-or have they?

This sort of situation epitomises the depths that our negotiators in Brussels have sunk to, and its up to the general public and the electrically orientated press in this country to say STOP -ENOUGH-we know what's right, and we refuse to be browbeaten into accepting an expen. sive second-rate article.

## Tiny Tele!

FOR the past four or five years, rumours have been flying around thick and fast concerning the imminent appearance of a pocket sized TV. It's true to say that a couple were produced and shown to the Press, but were never marketed due to some fundamental design fault. There were also more, we are led to believe, that were never revealed to the Press, but were sent back to the "Drawing board" for a further re-think. However, Sin-clair-possibly one of the leading exponents of the miniature electronics industry in this country, are now producing a truly tiny TV, which really is "Pocket" size -well large pocket size anyway! -and are claiming a World first.

Called the Microvision, it's just $6 \times 4 \times 1{ }^{1}{ }_{2}$ in with a 2 in screen, and is claimed to work just about anywhere in the Western World. It can tune in to any UHF or

VHF channel in bands $1,3,4$ or 5 and consume less power than a large pen torch bulb-to be exact 750 mW from 4 AA nickel cadmium rechargable batteries. When fully charged, life is about 4 hours, with a 14 hour recharging cycle. Audio output is 50 mW , while EHT to the tiny tube is 2 kV .
The Microvision is the result, say Sinclair, of 12 years research and some $£ 500,000$ spent in development work alone. Readers may also remember that the NEB granted a large sum of money to Sinclair, which really does prove that some Civil Servants know a good thing when they see it!

Manufactured at Sinclair's Headquarters at St Ives, the Microvision is priced at $\$ 300$ in the USA or $£ 175 \cdot 00$ plus VAT in the UK.

Sinclair Radionics Ltd., London Road, St Ives, Huntingdon, Cambs PE17 4HJ. Tel: (St Ives) 048064646.


## Blex Marshall

IT IS with regret that we have to announce the death of Alex Marshall, founder of A. Marshall (London) Ltd. of Cricklewood Broadway, London, after a short illness. Mrs. Marshall is continuing the business with the same management staff.

## Sution \& Cheam RS

THE ANNUAL Dinner Dance, will take place on Saturday 19 March at the Woodstock Hotel, Morden when the guest of honour will be the President of the RSGB, the Lord Wallace of Coslany. Details from Bob Tillin G3MES, 11 Great Ellshams, Banstead, Surrey or telephone Burgh Heath 56095.

## CONSTRUCTION

You may wish to construct the whole unit at one go and then test it, but you can run the risk of being faced with a totally baffling case of absolute silence at the end! The method to be described is a "build and test as-you-go" procedure which to some extent avoids this. Construction cannot be described as elementary, but neither is it particularly difficult. Be methodical in your approaoh and do not try to rush. First to be assembled is PCB1 according to the plan given in Fig. 5b. If you have not attempted PCB construction before, do not be afraid of making a start now. Follow the usual rules when mounting components; semiconductor connections are given in the drawings. With the board drilled as shown, the LP1186 and BLR3107 modules will only fit in one way round, as will L2. CF1 and CF2 may be mounted either way round. The relay may present a slight problem.


Fig. 6. Diagram of the reed relay assembly as used by the author.
Fig. 6 shows how the reed was mounted in the coil in the prototype, when a fairly standard type was used. If the relay used has different base connections, either it or the board will have to be suitably altered. Fig. 7 shows the heatsink for Trl, which is mounted vertically as indicated. Note that
to avoid the tuner current passing through the fixing screws of the heatsink a separate lead is soldered between the heatsink and the PCB. Apart from the power input, aerial and output leads, and the tuning voltage lead to the tunerhead, leave all flying connections off at this stage, which makes initial testing considerably easier. Finally, check all wiring thoroughly against the diagrams.


Fig. 7. Details of the heatsink for transistor Tr1, which is bolted on direct without any insulating kit. Avoid shorting the heatsink, which is at +25 V , to earth.

Figs. 8 and 9 show the layouts for PCB2 and PCB3. Notice that PCB3 is not drawn full length; to each end of what is shown in the diagram should be added another $13_{4}$ in of unetched board, making a total board length of 11 in . On these unetched portions holes should be drilled to take the mains switch and fine tune control (both of which are bolted to PCB3). Brackets may be bolted on for fixing the completed assembly of PCB2 and PCB3 to the chassis bottom. The positions for these holes can be calculated from Fig. 10.

When mounting the LEDs observe that they lie on the copper side of PCB3, with their leads taken in a loop and back through the second set of holes provided on the board, for soldering. C45 to C52 must be bent flat on the board in order to leave clearance for mounting PCB2.




PCB2 itself is relatively straightforward to construct and assemble; use screened lead where indicated; this is important as wires carrying the tuning voltage are particularly likely to introduce hum into the system, which then appears at the output.
Next construct the first part of the chassis. It will be seen from Fig. 11 that the metalwork is made of 18SWG aluminium in two L-shaped pieces. One forms the back and bottom, the other the front panel with a 25 mm (lin) flange for attaching it to the bottom. It is the first-mentioned piece which must be made now; cut, bend and drill as appropriate, making any alterations as necessary for the components actually used. For accurate drilling it is probably best to locate the positions of the holes for PCB1 by using the board as a template. Any treatment intended for the back panel such as labelling should be done now. The prototype was sprayed white, the lettering put on and then a very thin protective coat of polyurethane varnish sprayed on top. Allow adequate time for the paint to thoroughly harden off.


Fig. 12: Method of connecting the four 1N4001 diodes to the secondary tags of the mains transformer to form the bridge rectifier.

Now mount the switch S 2 , the sockets, the mains transformer and capacitor Cl . If the recommended transformer is used, its windings should be wired in parallel and in phase Fig 12, and the four secondary tags may conveniently be used for soldering the diodes D11 to D14 into the circuit. Temporarily connect a mains supply direct to the transformer, connect C1 in circuit, and ensure that this unregulated supply gives about 35 V . If all is well, PCB1 may be mounted on the chassis floor and the aerial, output, and power supply connections wired up.

## PRELIMINARY TESTING

Note that C57 and C58 are mounted directiy on the DIN socket, not on PCB1. Wire in S2 but during testing leave it switched off (open circuit) until testing the mute circuitry. Leave the Cl end of the wire link on PCB1 between C1 and the rest of the board circuitry unconnected at this stage. Switch on and if all is well, Tr1, ICl and IC2 should all remain cold; VR1 and VR2 are adjusted to give 12 V at the output of the main supply regulator, and $17 \cdot 5 \mathrm{~V}$ from the tuning voltage regulator. Then connect a milli-ammeter across the broken link mentioned above, thus delivering power to the boarci. About 70 mA should flow and the supply rail remain at 12 V . See that none of the ICs or transistors are hot; all the ICs and Tri will be warm under normal
conditions. When satisfied, the link may be soldered in place.

Plug in an aerial and amplifier, and feed a tuning voltage to the tunerhead using the $17 \cdot 5 \mathrm{~V}$ supply and a $100 \mathrm{k} \Omega$ potentiometer. With L2 adjusted so that the top of the core is one or two turns below the top of the screening can it should be possible to tune in a station. Adjust the core for maximum audio level (corresponding to maximum inter-station noise). Find a stereo transmission and adjust VR8 so that the decoder separates the two channels: if necessary an LED may be connected in place of LED9 between R55 and the 12 V rail which will then illuminate when a station is being received in stereo. Check the voltages provided in Table 1. If at any stage it is suspected that something may be wrong, switch off immediately and check the circuitry.

TABLE 1

| Test | Voits |
| :--- | :--- |
| Unregulated supply when loaded | 22 to 26 |
| Main supply | 12 |
| Tuning supply | 17.5 |
| Decoder supply (top of C40) | 10.5 |
| Detector supply (top of C23) | 10 |
| IF amp supply (top of C19) | 9.5 |
| Tunerhead supply (top of C18) | 7.8 |
| IC4 reference volts (pin 10;1C4) | 5.6 |
|  |  |

(All measurements made with a $20 \mathrm{k} \Omega / \mathrm{V}$ voltmeter on 5 or 25 V range as applicable).

Now attach all the leads to PCB1 not previously soldered in place, and connect all the wires leading between PCB2 and PCB3, and those between these two boards and PCB1 (the $100 \mathrm{k} \Omega$ potentiometer used earlier is not now needed). Bolt PCB2 and PCB3 together and mount the assembly on the bottom of the chassis using two small brackets as indicated in Fig. 11. The four rubber feet should also be attached, and the mains switch S1 wired into circuit. The copper side of PCB3 is painted matt black all round the touch buttons (the positions of which are indicated in Fig. 10) so that above, below, and at the ends of the row of buttons a ${ }^{1} 4$ in border of black exists. The copper-less lines in between the electrodes on each button should also be painted.

The stations should be labelled, from left to right, $2,3,4, L, L, L, L, L$, (three BBC national stations and five local). This may be done either on the central electrodes (in which case a small covering of varnish should be placed over the letters if 'magic letters' are used, but not so as to cover the whole electrode which would then insulate it), or on the front panel underneath the buttons.

## SELECTOR TESTING

Now comes the station selector testing. Switch on once more and ensure that the two supply rails remain at their correct voltages (an indication that the circuitry is not taking too high a current). One of the station LEDs should be on, and touching other buttons should select the corresponding stations; try adjusting the tuning presets to see that they all tune the receiver satisfactorily. When a stereo broadcast is being received the stereo beacon should be illuminated.

To set up the switching levels of the signal strength and tuning indicator LEDs it is necessary to peak up the tuning of L2. Detune the receiver so that a station is just distorted, and connect a link across the reed contacts of relay, RLA. If the AFC is working, the station should return to tune. Connect a voltmeter across the reed contacts; set the fine tune control to its midway position and adjust a tuning preset so that a station is in tune (when no voltage should be read). Detuning the receiver with the fine tune control should give equal but opposite voltage variations when the control is moved to equal positions either side of the midway position. If not, adjust L2 and re-tune the receiver such that this is so. It should be possible to do this with variations up to about 3 V , which will mean that the system is sufficiently well adjusted for use.

## AFC ADJUSTMENTS

Lastly the AFC override circuitry and the DIL socket defeat functions must be tested. During the above work, it will be found that when the tuning LEDs go off, the output is muted. Now insert a diode across pins 2 and 15 of DIL1, in the orientation shown in Fig. 4. The mute function should now be inoperative on Radio 2; that is, instead of the output being quiet when the tuning lights are extinguished, the station will gradually fade into a loud background hiss as the receiver is detuned. The station should also be permanently in mono. The same applies to diodes inserted across corresponding pairs of pins for other stations. Now similarly insert a diode between pins 1 and 16 on DIL1. Again select Radio 2, where the AFC should now be operative,


View down on to the Touch Tuner wilh mains transformer and smoothing capacitor to the left, both bolted direct to the chassis. The Tri heatsink is at extreme left of main PCB while PCBs 2 and 3 are flted behind the front panel. The longer of these iwo boards carries the fine tuning control and mains swilch. The shorter board carries the eight station tuning potenllometers which can be clearly seen.

Now place the voltmeter across R8. Turn VR5 fully anticlockwise and adjust VR3 and VR4 so that LED14 and LED13 switch on and off at voltages about 0.4 V below and above the in-tune voltage respectively. Leaving the receiver exactly in tune, place the voltmeter on pin 13 of IC4, and adjust VR5, VR6 and VR7 so that LEDs 13 and 14 switch on at 2 V, LED12 at 3 V , and LED11 at $4 \cdot 5 \mathrm{~V}$. Alteration of the voltage on pin 13 for this purpose may be done by de-tuning the receiver, but the best method is to remove the aerial and then slowly bring it up to the centre pin on the socket. When the receiver is tuned to a strong signal and the presets correctly adjusted the LEDs should progressively switch on as the signal strength is increased by the method outlined.

It is emphasised that all the LED adjustment should be done when the tuner is warmed up as the level indioator circuits are rather temperaturesensitive. Under normal operating circumstances the room temperature should not vary too much, but if it were to do so the voltages at which the LEDs switch on would vary considerably. With the signal level indicators set to the above values, when LED 12 is on the signal is strong enough for reasonable mono reception, and when both LED11 and LED12 are on then noise-free stereo reception should be possible.
signified by the fact that the sound remains clear and undistorted and the tuning indicators permanently on, over a considerable frequency range. At the extremes of this range the AFC will suddenly cease to operate and the station will be lost, denoted by the extinguishing of the tuning and signal strength LEDs. Try holding your finger on the Radio 2 button; it will need a firmer touch than when selecting a station, but the AFC should be rendered inoperative by this action, and the AFC beacon will go out. This permits accurate tuning of the station.

The same can be done for all the other stations by inserting a diode across the appropriate pins. Switching S2 to the off position will permanently defeat the mute function. This completes the electronics of the tuner and you can now insert diodes as required, and tune in the stations. The in-tune position is, of course, indicated by both tuning LEDs being illuminated. Although a simple piece of wire will suffice as an aerial in many cases, the benefits of a better arrangement are considerable when trying to receive high quality stereo broadcasts or signals from distant transmitters.

The last part of the work is to construct the front panel Fig. 10. As stated before, this is an L-shaped piece of aluminium (Fig. 11). Drill holes on the flange so that it may be fixed to the chassis bottom with the same bolts as are used to keep PCB2, PCB3
and the front rubber feet in place. Great care must be taken when cutting and drilling the front itself; there is considerable use of straight lines and symmetrical patterns which, if not strictly adhered to will mean that the panel will not coincide with the components on PCB3, and the appearance will be disastrous! Although positions are given in Fig. 10 for drilling the holes and cutting the slot it is strongly recommended that each hole is positioned with reference to the reader's version of PCB3. This means placing the panel in position against the LEDs (it may be necessary to remove the mains switch and fine tune control) and marking where the holes should be drilled on the inside of the panel.

Now drill each LED hole, one at a time; drill them only to 0.2 in diameter and file out the hole to take the black clip. This way ensures that straight lines of LEDs are obtained. Similarly drill the holes for the spindles of the controls, and cut out the slot. Any painting and lettering intended should be done at this stage, removing the LED clips to do so, of course.

The prototype was treated in the same way on the front and back panels, as has already been described. When the paint has hardened, glue to the inside of the front panel, all round the slot, some ${ }^{1}$ in square balsawood rod, painted matt black. This will hide undesirable parts of the printed circuit from view and prevent reflections from the LEDs on the buttons.

A three-sided wooden sleeve was made for the tuner from teak veneered plywood. Simple tack and glue joints were used, with wooden fillets for strengthening the corners. Small $90^{\circ}$ brackets were bolted to the inside of the sides so that when the sleeve was fitted on the chassis these brackets lay on the floor of the chassis. Self-tapping screws, put through holes in the floor from the underside, then held the brackets rigidly to the chassis.

## NOTES

Clearly there are faults that could be encountered during construction; the most likely cause is a wiring error. If the receiver has a background hum on some or all stations check the decoupling on the tuning voltage lines, both into IC6 and IC7, and from the presets which tune the stations. Also try reducing the main supply voltage a little; the system will work with as little as 9 V .

It may be that some of the stations cannot be selected. or that the receiver sticks on one position and pushing other buttons has no effect. Moisture on the electrodes could cause the latter fault, although the former is more likely to be a wiring error. It should never be possible to simultaneously have two stations selected (unless you hold fingers on two buttons). When first switching on it may be that although a station light is illuminated, no sound is heard. This is normal for these ICs and touching that button will immediately bring the station into tune.

Please note the following amendments to Fig. 4 in Part 1.

1. The common line at the bottom of R78 to R85 and C45 to C52 should be shown as going to +12 V .
2. LED20, in the collector line of $\operatorname{Tr} 26$, should be marked LEDIO.
3. Capacitor C53 is not polarised and should not be shown as an electrolytic.

## - THE DECCA 80 CHASSIS

The Decca 80 chassis, released in early 1976, is representative of the latest approach to colour set design, featuring an in-line gun c.r.t. and extensive use of new i.c.s. Barry Pamplin describes basic circuit operation, fault diagnosis procedures and common faults. The first of two articles.

## CRT BOOSTER

A c.r.t. booster can save money and do wonders for a set displaying a dull picture. This design, by Andy Denham, can be used with both colour and monochrome tubes and can be built up using components from the spares box. Boosting should give a tube at least six months' extra life.

## HAVE YOU NOTICED . . . ?

What's Les up to this time? His latest discourse on servicing experiences describes various failure patterns he has encountered time and time again.

## SCOPE TUBES

Oscilloscope tubes are constructed to meet quite different requirements from the normal receiver display tube. This can be confusing to anyone selecting a tube to build an oscilloscope around. 'Phosphor' describes basic scope tubes and the features they offer.

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## in UHF modulutar E.TRUNDLE

THE modulator to be described is based on the standard Mullard varicap tuner type ELC1043 or ELC1043/05. These are currently being advertised in this and other magazines.

The basic idea is to isolate the RF amplifier and self-oscillating mixer sections after which the oscillator is tuned down so that it comes into line with the RF amplifier. The oscillator output is then applied to the redundant RF input so that the RF stage amplifies the oscillator's signal. The video input is applied to the AGC line to vary the gain of the RF amplifier, the net result being a negativelymodulated RF signal at UHF, whose frequency may


Fig. 2. The location of R2 and the three tuning presets, R5, R11 and R13.


Fig. 1. Circuit of the ELC1043 after modification. The ELC1043105 is similar apart from transistor types. The shaded areas indicate breaks in the P.C. tracks. R31 is added in series with 218 to provide the oscillator output tapping point.
be set in the usual way by adjusting the bias on the tuning diodes from an external source. The RF output is several millivolts, sufficient to drive several sets if required, and may require attenuation if a signal set is used.

Unlike some commercial types, few spurious outputs are generated and such problems as microphony and drift are virtually eliminated.

## TUNER MODIFICATIONS

The tuner circuit, after modification, is shown in Fig. 1. The bandwidth of the AGC line is first increased by replacing the ceramic fecdthrough capacitor C4 by one of lower capacity, eg a ferrite bead or sleeved wire, and the original R2 (470s2) by R32 (47 ) , see Fig. 2 for location.


Fig. 3. Layout of the printed boards and modification delails for the ELC1043l05 (lop) and ELC1043 (bollom). In each case the left-hand drawing shows the print pattern before modification. The right-hand drawing shows track and components to be removed in broken lines, links and components to be added In full lines. Links should be made in 16 or 18 swg tinned copper wire, formed to stand clear of the board. External connections are identical for both versions.

The rest of the modifications are to the print pattern inside the tuner and are illustrated in Fig. 3 for each type. The link between L11 and L12 is severed, and an earth bridge soldered across the gap. The RF output is taken out by a thin screened lead from the "top" end of L11 through the vacant hole adjacent to pin 5.

One end of L18 is isolated from earth and a $10 \Omega$ carbon resistor, R31, fitted across the break. From the same point a screened lead is routed to L1, the original aerial input. The existing aerial input tag should be removed.
The lecher line, L14, is not required for our purpose, and the control potential from R13 is diverted to D4 by disconnecting the lower end of R21 from the junction of R20/R24 and linking it to R13 slider, which must be isolated from R14.

Finally, D11, VR1 and R30 are fitted inside the tuner on the print side, being suspended in the wiring, as shown in Fig. 3. The now redundant IF output pin (no. 10) is cut inside the tuner and used to anchor C34, the video coupler, which is mounted externally.

## POWER SUPPLIES

The modulator requires a 12 V line, positive of chassis, which must be capable of supplying 14 mA with a peak-to-peak ripple not exceeding 50 mV . A suitable circuit is given in Fig. 4, using a subminiature mains transformer. L25 is a UHF choke mounted on the tuner between pins $4 \& 8$, and consisting of a few turns of ordinary plastic insulated wire. F1 is important from the safety angle and should not be omitted.

A further supply line is necessary for tuning purposes. If a fixed-frequency UHF output is required, the arrangement of Fig. 5 may be adopted, VR2 taking the form of a slider or rotary potentio-


Fig. 4. The Power Supply and Input Bias circuits.

Fig. 5. The tuning circuit for a preset spot frequency output.

To D6 anode


Fig. 6. The circuit for continuous coverage of Band IV and V.
meter mounted on the tuner body. This will give pre-set coverage up to about channel 44. The more elaborate circuit of Fig. 6 utilises a voltage tripler and an IC to give a stabilised 32 V line derived from one-half of the transformer secondary. This offers coverage of the UHF band up to channel 68 but at greater expense.

A suitable arrangement of the components required for the power supply and the variable, total coverage, tuning circuit of Fig. 6 is shown as Fig. 7. This drawing also shows the arrangement of the tuner and the extra board in the box.

## components list

## General

Varicap tuner type ELC1043 or ELC1043/05. Aluminium box, $102 \times 102 \times 38 \mathrm{~mm}\left(4 \times 4 \times 1 \frac{1}{2} \mathrm{ins}\right)$ with lid. 2 surface mounting coaxial sockets. Knob for VR3 (when used).

Power supply and input bias cet.
R27 $120 \Omega 5 \%$, $\frac{1}{2} W$
R29 82 5 \% , 交W
R30 $4 \cdot 7 \mathrm{k} \Omega 5 \%$, $\frac{1}{2} W$
VR1 $1 \mathrm{k} \Omega$, min. skel. horz.
C28 $220 \mu \mathrm{~F}, 25 \mathrm{~V}$
C29 $220 \mu \mathrm{~F}, 25 \mathrm{~V}$
C34 $100 \mu \mathrm{~F}, 25 \mathrm{~V}$
D5 BA145 or BA148
D6\} or BY206
D10 BZY88C12V
D11 BZY88C6V2
$\mathrm{T} 1,12-012 \mathrm{~V}$ at 50 mA or 100 mA subminiature mains transformer. F1, 80 mA , antisurge fuse ( 20 mm ) and holder. L25, home made inductor (see text). Mounting board.

```
Tuning circult (Fig. 5)
    R33 20ks 5%, iW
    VR2 22k\Omega min. skel. horz.
    C35 1 LF F,35V, tant.
```

Tuning circuit (Fig. 6)
R28 $5.6 \mathrm{k} \cap 5 \%$, 士W
VR2 47 kn , min. skel. horz.
VR3 $100 \mathrm{k} \Omega$, carbon, linear
C30 $22 \mu \mathrm{~F}, 35 \mathrm{~V}$
C31 $22 \mu \mathrm{~F}, 35 \mathrm{~V}$
C32 $47 \mu \mathrm{~F}, 63 \mathrm{~V}$
C33 $22 \mu \mathrm{~F}, 63 \mathrm{~V}$
D7 BA 145 or
D9 BY206
IC1 TAA550 or ZTK33B or SN76550

(a)


Fig. 7. A diagramatic representation (above) of the extra board, with wiring for the components of the power supply and continuous coverage tuning circuit shown. A photograqh of the inside of the prototype (below) shows that although the extra components are not too cramped there is not much surplus space avallable.


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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 50 | $0 \cdot 20{ }^{\circ}$ | 0.25 | 0.35 | 0.32 | 0.41 | 0.42 | 0.47 | 0.98 |
| 100 | $0.25{ }^{\text {* }}$ | 0.25 | 0.40 | 0. 37 | 0.47 | 0.48 | 0.54 | 102 |
| 200 | $0.27 *$ | 0.35 | 0.45 | 0.40 | 0.58 | 0.60 | 0-68 | 1-14 |
| 400 | $0.30 *$ | 0.40 | 0.50 | 0.45 | 0.87 | 0.88 | 0-98 | 1.40 |
| 800 | - | 0.65 | 0.70 | - | 1.09 | $1 \cdot 19$ | 1.25 | 1-80 |

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# S-PeGnolegy 



THE majority of "lamp flashing" circuits seen in the popular constructional journals seem to employ a two-transistor multivibrator circuit with a bulb as one of the collector loads. In this month's S-Dec circuit I've tried to reproduce similar results, but with different circuitry.

The final arrangement is shown in Fig. 1. Readers who have built the previous projects in this series will see most of the components used again in this circuit.

## Circuil details

Note that $\operatorname{Tr} 1$ is a BC109, a NPN type transistor, whereas $\operatorname{Tr} 2$ is a AC126, a PNP type. These two semiconductors are connected in a complementary arrangement and are DC coupled, i.e. the connection


Fig. 1 ; Clicult diagram of the Simple Stroboscope. Varying the rate of flash can be achieved by experimenting with the values of C1 and R3. The lower llnk should be connected between 35 and 66.
from the collector of $\operatorname{Tr} 1$ to the base of $\operatorname{Tr} 2$ is direct and not through any capacitor which, of course, would block the DC. Transistor Tr2 derives its base bias from a resistive divider chain formed by R4 and $\operatorname{Tr} 1$. The load for $\operatorname{Tr} 2$ is the lamp LP1 connected in its collector lead. The circuitry around Tr 1 seems quite conventional, too. A base bias from the resistor chain (R3, VR1, R1, R2) and R4 as a collector load.

Then there's the capacitor C1. This is a feedback component which applies the output signal back to the input or base of $\operatorname{Tr} 1$ and this is why the whole circuit oscillates.

The oscillator is of the " RC " type, meaning that its frequency of operation (or the rate at which the bulb flashes on and off) is determined and controlled by resistance and capacitance values. Since there is only one capactor in the circuit then this takes care of the " C ". 'The resistance variation is afforded by the potentiometer VR1 allowing the flashing rate to be varied over a set range. The other resistance in series with VR1 also have an effect on the flashing rate.

## Construction

To build the circuit, simply plug the components (via their own leads) into the S-DeC holes indicated on the diagram of Fig. 1. The lamp, for example, is
you will need . . . .

|  |  |
| :--- | :--- |
| R1 | $5 \cdot 6 \mathrm{k} \Omega$ |
| R2 | $5 \cdot 6 \mathrm{k} \Omega$ |
| R3 | $39 \mathrm{k} \Omega$ |
| R4 1 kS | Tri BC109 or 2 Ni 613 |
| VR1 $50 \mathrm{k} \Omega$ pot | B1 12 V battery |
| C1 $4.7 \mu \mathrm{~F} 12 \mathrm{~V}$ | solid core wire for wiring |
|  | One S-DeC |

plugged into holes 59 and 69 while R2 is plugged into S-DeC holes 22 and 32. The three lines marked "links" in Fig. 1 are simply short lengths of wire which have their ends plugged into the S-DeC holes as marked. Figure 2 shows a full size drawing of an actual S-DeC and the holes and component locations can be clearly seen. Be careful when you plug in $\operatorname{Tr} 2$, remember that it is a PNP device unlike the BC109, and has a small spot or mark on its case signifying that the lead nearest to it is the collector.

## Component trials

One of the great advantages of using an S-DeC is the ease and simplicity in experimenting with different devices and alternative values for components. Constructors might like to conduct their own experiments to see if they agree with my own findings.

For example: why is $\mathrm{R} 339 \mathrm{k} \Omega$ ? Why not $47 \mathrm{k} \Omega$ or l0k $\Omega$ ? When the circuit was first "plugged in" on the S-DeC, R3 was omitted. After all, if one can vary VR4 one could obtain a variable resistance.

The first attempt to get the circuit to oscillate used a $10 \mathrm{k} \Omega$ potentiometer for VR1. The result was that the BC109 got very hot very quickly! Look at Fig. 1 and you'll see why. If VRl is at zero resistance, then the base of Tr is connected via Rl to the positive 12V rail (I didn't have R 3 in circuit) so the transistor was biased very hard "on". But even with $10 \mathrm{k} \Omega$ in circuit (i.e. maximum setting of VR1) the BCl09 was still hot and the flashing rate was not very good although the circuit did work. Substituting a $50 \mathrm{k} \Omega$ pot gave some improvement but again, if the pot was set to minimum resistance, the base of Tr l would be connected via R1 to the 12 V positive rail. Clearly a "safety" resistor was needed in series with the pot to ensure that there was always some sensible resistance value between Tr 1 base and the 12 V rail.


Rearview photograph of the S-DeC showing the simple nature of the circuit. It's advisable to use holders for the transistors, as their leads tend to be on the short side.

Various values were tried for R3 but the optimum value was found to be $39 \mathrm{k} \Omega$.

Changing the value of the capacitor in this circuit did not prove a good move. Using a $10 \mu \mathrm{~F}$ component reduced the flashing rate, and swinging VR1 from maximum to minimum gave far less variation in the rate than with a $4 \cdot 7 \mu \mathrm{~F}$ capacitor for Cl .
All values given in Fig. 1 were found to be optimum although there is no reason why readers should not experiment for themselves with other values. The 2 N 1613 used in earlier S-DeC projects in this series can be substituted for the BC109 with very little effect on the overall performance. (Lead connections the same.)


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## Frozen motion

If this circuit is to be used as a miniature stroboscope, then the bulb LP1 should be put into some kind of reflector. This will increase the intensity. A cycle lamp reflector is useful but a simple "cup" made from aluminium baking foil is also effective. By shining the light onto something in motion, the motion can be apparently stopped or frozen provided the speed of the movement is within the range of the strobing light or a certain mathematical relationship to it.

If you set the flasher up and point it at a tap in a darkened kitchen, then by adjustment of the tap and the flashing rate, the apparently solid stream of water can be broken up into many tiny individual droplets all suspended in space.

A further use for this month's circuit would be to take it with you to a party. Here it could be used as a pocket disco light. Lay in flat with the bulb and reflector pointing at the ceiling ( coz ceilings are usually white and they reflect well). The result is a kind of freezing of motion as people "dance".

## A warning

A word of warning about this stroboscope circuit. Under no circumstances use it near anyone who is subject to epilepsy or fits. It will probably trigger them off and it could be very dangerous.

Be careful about looking into the strobe, especially in a darkened room. At some frequencies strange "hallucinations" can be caused in some people. It may be that the flashing rate is, at a particular setting of VR1, synchronised to the brain rhythms of the viewer.

Most of the projects in this series have been aimed at amusement plus, hopefully, some more serious suggested uses. In next month's Practical Wireless the S-DeC circuit will be one which will be of use to everyone and anyone who builds or constructs. An absolute must for any workbench which will save time and money. It's simple, easy to use, very easy to construct and it's called, simply . well, bless me, I've run out of space-but I will tell you all about that circuit-in the next issue.

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(1) P83l, col 2, line 7 -for 'sockets N, K, J, J' read 'sockets $\mathrm{A}, \mathrm{B}, \mathrm{C}, \mathrm{D}^{\prime}$
(2) P832, col 1 , line 3 -delete the figure ' 5 '.

## AUDIBLE CONTINUITY TESTER February 1977

Note that R4 in the components list should be $180 \mathrm{k} \Omega$ as in the circuit diagram and not $100 \mathrm{k} \Omega$.

## CHROMACHASE 4.

## December 1976/January 1977

1. A fault in the diode matrix prevents operation in the 'forward' modes.
To correct it, remove the cathode (ring end) of D8 and stand the diode vertically. Insert the anode of an extra diode into the hole left by D8. Stand this diode vertically also. Remove the wire from Sla to edge pin 15. Solder a new wire from Sla to both cathodes of the vertical diodes. The diode should be the same type as D8.
To correct the circuit diagram (page 676) replace the direct link between edge pin 15 and the Reverse Stop line by a diode symbol in the same direction as D8.
2. The rotary switch for 'Program' works in reverse to the front panel. To correct, reverse the wiring connections to the switch.
3. A drawing error has been made in the circuit on pages 677. The symbol for $\operatorname{Tr} 10$ is shown as NPN when it should be PNP. Provided a PNP transistor is used the drawing error has no effect. To correct, draw the arrowhead inwards on the emitter instead of outwards.
4. A possible fault can occur in that lamps lock on and the display isn't as set. This would be due to sharp pulses from the triggering transformers, T2 to T5, being picked-up on the leads and wiring of IC4. To correct it, add decoupling capacitors ( $0 \cdot 1, \mathrm{~F}$ ) between ground and the following points,
A. IC4 pin 9
B. IC4 pin 10
C. IC4 pin 11
D. IC4 pin 16 J
(The circuit board already has holes for these components. The prototypes functioned without them so they were not included on the components list.)
E. If yet more decoupling is required, add $0 \cdot 1$, $F$ 's from ground to IC4 pins 4,5 and 6.



## PRAGTIGAL WIBELESS



## Here's another!

Have you wondered about the "how" and "why" of semiconductors or what those "cheap" transistors are good for?
Some of the answers are given in our second Info-Card which introduces the Semiconductor and shows generalised curves as well as classifying some 70 commonly available transistors.

## PLUSTHE



This new locator is a significant advance in the tried, tested and approved range of Beat Frequency Detectors.
Externally it offers a distinctive shape and exchangable, fully waterproof search coils. Internally the circuitry allows detection to a convenient depth in most non-metallic materials with switched sensitivity and meter plus earphone or speaker monitoring.

THIS power unit has been designed for general purpose, bench use. It is based on the Integrated Circuit regulator type 723. Whilst this regulator is suitable for a wide variety of stabilised circuits this is probably the most common.

The principles of the regulator are the same as those for discrete component devices, i.e. a portion of the output voltage is compared with a known, stable, reference and the difference used to control the output. If the output rises for any reason the positive variation decreases the base current in a

## POWIR S

series control transistor which, in turn, decreases the emitter (output) current and the output voltage falls until it becomes that originally set. Similarly,


1. Current sense
2. Inv. input
3. Non-Inv input
4. Vref
5. OV
6. D.C. out
7. Ve
8. D.C. in
9. Freq. comp.
10. Current limit
To5 style
Viewed from above


Fig. 2. The clrcult dlagram of one supply. The second supply is identical and the components have been numbered in the serles R101, C101, where R1 and R101 perform the same functlons In their respective clrcults. Note that Tr1 should be 2N3055.


## T.J. JOHNSON

the circuit increases the emitter current if the output voltage falls below that set. This correction occurs very fast and, to all intents and purposes, the output stays constant.

## components list

## Resistors

R1 and R101
R2 and R102
R3 and R103
R4 and R104
R5 and R105
R6
VR1 and VR101 $50 \Omega$ Wire Wound VR2 and VR102 $5 \mathrm{k} \Omega \mathrm{lin}$.

## Capacitors

C1 and C101 C2 and C102 C3 and C103 C4 and C104
$820 \Omega 2 \%$ W
$390 \Omega 5 \%$ 古 $W$
$470 \Omega 5 \% \frac{1}{2} W$
$470 \Omega 5 \% \frac{1}{2} W$
$0.5 \Omega 5 \%$ iW ( $2 \times 1 \Omega$ in parallet for each)
To suit meter used (approx $10 \%$ of the meter resistance)
$200 \mu$ F. 50 V
$0 \cdot 1 \mu \mathrm{~F}$ Polyester
$100 \mu \mathrm{~F} 50 \mathrm{~V}$
$4.7 \mu \mathrm{~F} 50 \mathrm{~V}$
Semiconductors

IC1 and IC101 Tr1 and Tr101 D1 to D4 and D101 to D104

723 regulator, 14 pin DIL 2N3055, TO3 package

## 1N4001

## Miscellaneous

T1, mains transformer, two separate secondaries as required ( 16 V used in prototype). NE1, mains neon. S1, switch DPDT. S2, switch DPDT. S3, switch DPST. S4, switch DPST. S5, switch 4 pole 2 way rotary. S6, switch DPST. SK1 to SK5, terminals 4 mm (two red, two black and one green). M1, meter 0 to 20 V dc (type MR38p). M2 meter 0 to 100 mA dc (type MR38p). Four round knobs. One pointer knob. Heatsink for two transistors, $4 \frac{2}{2}$ " $x$ $4^{\prime \prime} \times 1^{\prime \prime}$ (Marston type C or similar). Stripboard, $0.1^{\prime \prime}$ track, $5^{\prime \prime} \times 37^{\prime \prime}$. Material for front plate and case, Covering for case. Two IC sockets, 14 pin DIL.

The pin comfiguration for both 14 pin DILs and 10 pin TO5 packages are given in Fig. 1. It should be noted that by aligning the pins on the TO5 package to have 6 in one line and 4 in the other the pins can be made to correspond to those of the DIL and can therefore be inserted into the same holes on the board.

The maximum current for the 723 is 150 mA and, since 1 A is required from the unit, external power transistors are used. The inbuilt protection of the 723 against short circuits, excessive power dissipation and overheating are still available when extra transistons are used.

## CIRCUIT OPERATION

The schematic circuit for one supply is given as Fig. 2. The second supply is identical in design and operation.

The mains supply is stepped down by T1 and the low voltage output is bridge rectified by diodes D1 to D4. Capacitor Cl provides a limited amount of smoothing to enable the 723 to acoept the voltage as its input.

Since the input is always higher than 12 V there is a reference voltage of a nominal $7 \cdot 15 \mathrm{~V}$ at pin 6 and this voltage is used for the stabilisation. However, since the unit is designed to operate down to approximately 2 V , the reference voltage has to be divided. R3, VR2 and R4 permit a range of voltages from 0.6 V to 5.6 V to be selected and applied to the non-inverting input, pin 5.

The output voltage is applied to a second, fixed, potential divider consisting of R1 and R2. The voltage developed across R2 is applied to the inverting input, pin 4.

Now, as the potential at pin 6 is varied, by adjusting VR2, the output voltage will vary in step with it.

The unit is required to produce up to 1 A output current'and as this is in excess of the IC capability a series regulator transistor, $\operatorname{Tr} 1$, is used. The DC

output at pin 10 is used to provide the base current for the transistor, the smoothed output of the bridge provides the collector voltage and current and the output is the emitter load.

The emitter current flows through VR1 and R5 but, since the sampling potentiometer is across the output load only, potentials dropped across VR2 and R5 do not affect regulation.
The 723 has a built in shut down transistor which comes into operation when the potential between pins 2 and 3 reaches 0.65 V . This voltage is developed across VR1 and R5 so the maximum current that can flow before the cut-out becomes operative is determined by the setting of VR1 (for large currents R5 is effective). The need for R5 is explained later.

Beoause of the limited smoothing of the supply voltage there is a certain amount of hum present on the output, unless C3 is fitted, despite the effect of the 723 .

Two meters have been used on the prototype, one voltmeter and one milliammeter. To enable the full range of output current to be used the milliammeter can be shunted, by S6, to read 1A. The value of R6 will depend on the resistance of the meter used and may need to be custom made from resistance wire in order to obtain a reasonably accurate scale. Both meters are switchable to read either output A or output B. S2 performs the switching for the voltmeter and 55 for the ammeter.

In addition to the mains ON/OFF switch, Sl , both supplies have individual switches, S3 and S4. The switching details are given as Fig. 3. Small dial meters have been used and this results in less than perfect readability, particularly on the ammeter, and if accurate settings of voltage or current are required a larger meter will be needed.

## CONSTRUCTION

The integrated circuits and the small components associated with them are mounted on a piece of stripboard. The size, cutting required, component location and orientation is given as Fig. 4. The two power transistors are mounted on the heatsink,

Fig. 3. The switching requirements to monitor both supplles by using only one voltmeter and one ammeter. The use of separate meters for each supply would reduce the switching signlficantly.
using insulated bushes and mica washers. The heatsink is fixed to the outside of the case to assist cooling.

The transformer and the board are bolted to the base of the case, which is chipboard to give the neoessary rigidity to the unit. The switches, meters and the variable controls are mounted on an aluminium front plate. The sides and top are plywood, preassembled to form an inverted ' $U$ '. The back is a combination of chipboard and plywood. To improve the appearance of the prototype the top, sides and back were covered with wood-patterned paper.
The major dimensions for the front panel are given as Fig. 5 and those for the remainder of the case as Fig. 6.

## SETTING-UP

Each supply should be switched on in turn and, with no load connected, the voltage control should be checked for the relevant supply. The output should indicate from approximately 2 V to 20 V .
Set the voltage control of one section to minimum and switch the ammeter to 1 A . Connect a high wattage resistor of about 20 ohms to the supply selected. Set the current control to minimum and the voltage control to about 10 V . No current should flow. Increase the current control and note that the current increases. The increase should continue until it reaches the expected value (with 10 V and $20 \Omega$ this would be 0.5 A ). When further increase in the control produces no corresponding increase in current the overload condition should have been reached. Apply a short circuit to the output. The current should not increase by an appreciable amount but the voltage should fall to nearly zero.
Remove the short circuit and repeat the test for the second supply.


Fig. 4. The cutting and component locations on the stripboard. The minimum number of breaks has been made. If nylon screws and nuts are used the breaks to isolate the fixing holes can be omitted.

Fig. 5. The layout of components on the front panel of the prototype. This layout is not critical and may be re-arranged to suit other cases or components.


Note: Cover is ${ }^{1} / \mathrm{g} "$ plywood constructed as a "U" shape with overall dimensions $10^{\prime \prime} \times 6 \frac{1}{2} \times 5 \frac{1}{2}{ }^{\prime \prime}$

Material: 5/8" chipboard $1 / 8^{\prime \prime}$ plywood

[^3]



## CMOS Iechnology

Complementary Metal Oxide Semiconductor Logic, to give it its full name is based on two types of MOSFETS, n-channel and p-channel enhancement devices. Each has four electrodes, Fig 1, the sub-


FIg. 1 : Drawing of a n-channel and $p$-channel complementary MOSFET device.
strate terminal is taken to one supply rail, $V_{D D}$ (positive rail) for p-channel transistors, and $\mathrm{V}_{\text {ss }}$ (-ve) for n-channel devices. The n-channel type can be thought analogous to an NPN transistor in the way it operates, although the gate takes negligible current and the source drain path when the transistor is switched fully on, appears as a resistance (about $500 \Omega$ ) rather than a voltage drop. For


Fig. 2: Transmission gate and Inverter of CMOS logic.
an n-channel transistor, the source is negative with respect to the drain, and conduction occurs when the gate is sufficiently more positive than the source. The p-channel devices require opposite polarities to those of the n-channel version.

Pairs of complementary MOSFETS are used to make up the two basic building blocks of CMOS logic, the transmission gate and the inverter, Fig 2. When an inverted is connected between the control pins, both transistors are either on or off and the transmission gate acts as an analogue or digital


Fig. 3: Combination of transmission gate and inverter, so forming a 2-input NAND gate.
switch capable of passing signals in both directions. The off/on resistance ratio of this solid state switch is better than 3000000:1.

Inverters, when connected in parallel/series combinations can be made to form gates as in Figs 3 and 4. However, the arrangements are asymmetrical,
hence the NAND has a greater pull up capability, while the converse holds for the NOR. If gates with more inputs are needed, the output impedance will rise, unless two extra inverters are added to the


Fig. 4: Similar combination as In Fig. 3, but forming a 2-input NOR gate.
output to restore symmetrical output characteristics. These extra inverters are standard on ICs with a B suffix.

## Circuit description

Reference to the circuit diagram in Fig 5 and the block diagram Fig 6, will show that the logic of the game splits into three main parts-the random time
delay monostable, control and scoring logic. The monostable comprises two NOR gates, capacitor C5 and resistors R7-R11. The time delay is determined by the combination of resistors selected by IC2 which, in turn, is controlled by IC1, a 7-bit binary counter (only 4 outputs are used). The latter is clocked by an oscillator formed by IC3c, IC6c, R4, R5 and C4. A logic 1 on the input to the NOR disables the clock.

The control logic consists basically of an RS latch IC4a and $b$, and various other gates and inverters which drive the indicator lamps, and score counter IC8. This is an up/down counter, the direction of counting being dependent on the output of the EXCLUSIVE OR gate, IC7d. A logic "l" on the up/ down input of the IC persuades it to count up, a logic " $O$ ", down. This up/down counting is the whole basis of the scoring system in Tug ' 0 ' War. The counter is clocked by inverters which are preceded by a pulse delayer R1 and Cl. A BCD output is also provided by the counter which drives IC9, a BCD to Decimal decoder which feeds the LEDs.

## Circuit operation

Play begins when both push buttons are depressed by the two players. This causes two logic " O "s to appear on the inputs of IC3, whose output then rises to a logic 1. This stops the counting action of ICl and so produces a random binary output on the four data lines linked to IC2. This IC contains four transmission gates which are wired across each of the resistors R7-R10. If a logic 1 is present on the output lines, the resistor associated with that bit of the counter will be shunted. The overall effect is to produce one of sixteen different resistance values


Genera/interior photograph of the Tug-O-War, showing the board layout, interboard wiring and switch positioning.

## * components list




Fig. 5 : Full circuit diagram of the Tug-O-War, giving details of the nine integrated circuits used in its construction. The'rope' is formed by LED's 1 to 9.
which controls the time constant of the monostable.
The positive pulse on the input of the monostable which also disables the oscillator means that the output of IC3 will go low. Since C5 is initially uncharged, the input of IC3b will go low also, taking its output high. The output of IC3d is held low during the timing cycle as the output of the monostable is connected to its other input. Meanwhile C5 charges at a rate dependent on the resistance chain, but when the voltage across it reaches the transfer voltage of the gate IC3b (approx 70 per cent of V), the output reverts to low level. This signal is buffered with two EXCLUSIVE OR gates and then inverted and delivered to the two "go" LEDs.

The RS latch connected to the two push buttons is set by the first player to release his button. The
other button will then have no effect. The two outputs are inverted, then compared with the monostable output by IC4c and d. Should the output be high when the latch is set, both inputs to the NAND in question will be high thus its output will go low and the respective "cheat" light will come on.

## Chealing

Two EXCLUSIVE OR gates IC7c and d, produce an output from the latch and the "cheat" outputs which causes the counter IC8, to count up or down. If neither player cheats, the outputs of IC4c and $d$ will not go low thus the output of IC7c will be low (an EXCLUSIVE OR gives a 1 output if its inputs differ). The level on the up/down input to the


Main PCB for this project, containing all components with the exception of IC8, IC9, LED's 1 to 9 and R6. Foil side above and component overlay below.



Fig. 6: Simplified block diagram of the logic used in this project.
counter will be high if the input to IC7d is high, and vice-versa. However, should there be a cheat in our midst and he releases his button before the GO signal, one of the inputs of IC7c will be high and the second EXCLUSIVE OR gate will invert its output, and reverse the direction of counting.

The clock pulse to the counter is initiated when one button is released, but it is not delivered until
the time interval set by Rl and Cl has elapsed (about $12 \mu s$ ). This short delay ensures that the counter is only clocked after all the other logic has settled. The inverters IC6d to f produce a fast rise time pulse for reliable clocking.


Fig. 7 : Dimensions for drilling the front panel which has all switches, and LED's for the 'rope' mounted on it.

Pressing the reset button sets the 'knot' in the centre of the rope of the LEDs. This is accomplished with the help of inputs to the counter which allow presetting of the counter to any required BCD code when the preset enable pin is taken high. In this case the button, when pressed, sets the output to 4 , which illuminates the middle LED.

## Construction

The unit was housed in a two-tone plastic verobox $150 \mathrm{~mm} \times 80 \mathrm{~mm} \times 50 \mathrm{~mm}$ which gives a very neat final appearance. The top should be drilled to take the LEDs and switches, hole positioning is shown in


[^4] between the top of the board and the LED's.

Fig. 7. When drilling the case, take care not to let the drill bit slip, as bitter experience has shown how easily an expensive plastic case can be ruined with little or no effort! The electronics is held on two printed circuit boards, to enable the components to be mounted in a reasonable space, and to reduce the amount of free wiring to the lid, see Figs 8 and 9.

Begin construction with the nine wire links, then go on to the passive components, flying leads and finally the ICs. The latter should be kept in their conductive foam until needed. Check the polarity of LEDs, tantalum capacitor and especially the ICs before soldering. The small red LEDs should be soldered as proud of the board as their leads will allow or they will not protrude through the lid. To this end the PCB is mounted with a nut between board and case in each corner so that the LEDs can be secured to the lid by means of the clips supplied.

When soldering the LEDs, beware of keeping the iron on them too long, as the red plastic is easily softened and then all that holds the leads on, is the light emitting junction itself.

Resistors R1 and R4 and capacitor Cl should be mounted flush with the board so that the battery will not foul the push button when it is installed. The push buttons S1 and S2 must be wired up correctly or the game may well become a push, rather than a tug ' $o$ ' war.

## Playing the game

After switching the game on, the score should be reset with the push button opposite the on-off switch. Both players should now press their push buttons whereupon the green 'go' LEDs will go out. After a random time interval which was from 2 to 10 s on the prototype, the LEDs will come on again. The first player to release his button after this signal, has the light moved one place closer to his end. A cheat who removes his finger before the signal will have the light moved one place closer to his opponent. Play then continues in this manner until the light reaches one end. The player at that end is declared the winner of the contest.

## Modifications

The game can be expanded for team use-the push buttons for each member of the team are wired in series. The fastest person in the two teams will then win the point but who it was . . . ?

A small problem encountered with the prototype was that the switches used were rather prone to contact bounce causing irregularities in the scoring, however one does get used to the feel of the buttons after a while. A worthwhile improvement would be to replace $S^{\prime} 1$ and S2 with touch switches etched from copper clad board offcuts, this would eliminate contact bounce and reduce strain on the finger whilst waiting for the 'go' signal. If this modification is carried out then R2 and R3 should be increased in value to about $3 \cdot 3 \mathrm{M} \Omega$; C 2 and C 3 may need to be reduced in value or omitted altogether to achieve the desired effect.

## THE in D. JONES 'SUREFIRE'




THIS project is aimed at those constructors who are unable to make an electronic circuit work first time. It is hoped that even the most hamfisted reader, with little trouble, should be able to build and operate this device.

The main function of the circuit, Fig. 1, is to do nothing, a very difficult achievement in a continually changing universe. The circuit layout has been designed so that it would take a very large number of mistakes for nothing not to happen. Should a constructor wish; peripheral devices can be added, but he is warned that too complicated a circuit may do something.


The battery was obtained from a transistor radio that did nothing when switched on; this is a good indication of the batteries condition. It should also do nothing when connected across a meter. But to make certain of its integrity a length of wire was connected between its terminals and left for a few hours. The two protection diodes D1/2 were taken from a burnt-out bridge rectifier, but new ones would work just as effectively.

The mechanical switch was included because it was felt that a touch switch would have increased the price of the project and may have caused something to happen contrary to the aim of this project. The potentiometers were added to fill two holes in the box used to house the circuit board, but having included them it became apparent that they were a 'must', especially when showing the finished product to admiring relatives.

VR1 acts as a Kenotron and causes all positrons present in the surrounding magnetic medium to undergo an adiabatic contraction. This contraction excites the ICW from VR2 causing the Lorenz transformations to be temporarily invalid within the immediate Hamaltonian space. The combined effect is that nothing is created and then passed on to the detector.

When the circuit is assembled and switched on nothing should happen. If, however, something does happen it is advised that the builder does not take up a career in electronics.


SINCE the introduction of gas and smoke detectors various circuits and constructional articles have been published, usually with features that may be redundant in a purely domestic unit. The majority of sensors available have unusual power requirements (typically $1 \cdot 2 \mathrm{~V}$ at 500 mA ), resulting in the need for specially wound transformers or, if portable, very short battery life. Since the introduction of the Type 812 sensor a very economic and easily constructed alarm system can be made without losing facilities or effectiveness and gaining, due to circuit simplicity, overall reliability. This article describes such a unit.
The circuit (Fig. 1) is basically a blocking oscillator formed by $\operatorname{Trl}$ and 2. In the presence of any de-oxidising agent such as smoke, gas, or petrol vapour the semiconductor coating of the sensor decreases in resistance which increases the bias applied to Tr base. This triggers the oscillator, producing both an audio alarm from the speaker and a direct intensity reading by the meter. VRl sets the sensitivity of the oscillator and may be set to trigger the audio alarm at any given gas/smoke intensity.

The supply requirement of the sensor circuit may be derived from the mains power supply (Fig. 2) or the unit can be made portable and a 12 V battery supply used. Diode Dl is included to give reverse polarity protection.

## Construction

The printed circuit board Fig. 3a should be assembled as shown (Fig. 3b). The front panel is then cut or drilled to suit the meter, speaker, LED and mains switch. Next the power supply fuseholder and sensor socket are installed in the case. The circuit board is then bolted directly on the meter terminals which establishes the connections to the meter. The unit may now be wired as shown, paying particular attention to the connection of the sensor socket terminals.

Set VR1 to mid-track and plug in the sensor unit. Now switch on, when the alarm should sound almost immediately and then rise in pitch. After five
seconds or so, as the sensor warms up, the pitch will decrease to a steady note, maintaining a meter reading, typically $100 \mu \mathrm{~A}$.

If a small amount of cigarette smoke or gas from a butane lighter is released near the sensor the meter reading will rise and the alarm will sound. The sensitivity control VRI of the audio alarm should be set so that the alarm is not triggered by



123 W419.


Fig. 1: top, is the circuit of the Alarm with Fig. 2, below, the mains power supply circuit. Both are incorporated in the Alarm shown in the heading photograph.


Fig. 3a, top, is the full size drawing of the PCB with the component layout, Fig. 3b, below.
any cigarette or cigar smoke that may be present normally. Usually VR1 is set at approx $75 \%$ rotation triggering the audio alarm at $200 \mu \mathrm{~A}(0 \cdot 2 \mathrm{~mA})$.

The unit described is suitable for continuous operation. Type 812 sensors have an approximate operational life of two years continuous use and may be periodically checked as previously described. Replacement is a mere formality, just plugging in a new sensor. When the sensor is nearing replacement time the reaction to gas and smoke will be sluggish.

## * components list

## Resistors

R1 $22 \mathrm{k} \Omega+1 \mathrm{~W}$ R2 $4 \cdot 7 \mathrm{k} \Omega \frac{1}{2} \mathrm{~W}$ VR1 $4 \cdot 7 \mathrm{k} \Omega$ submin. horizontal pot.

## Capacitors

C1 $10 \mu$ F 10 V C2 100 nF C3 $470 \mu \mathrm{~F} 25 \mathrm{~V}$

## Semiconductors

Tr1 BC148 Tr2 BC158 IC1 7805 regulator D1/2/3 1N4001 D4LED 0.2" and clip TGS812 Sensor with holder (Watford Electronics)

## Miscellaneous

Speaker $1 \frac{1}{2}{ }^{\prime \prime}$ dia. 4 to $8 \Omega$. Meter 1 mA FSO (Watford Electronics Type T24). Transtormer T1, $0-12 \mathrm{~V}$, $0-12 \mathrm{~V}$ 6VA total, 240 V primary. Fuse 500 mA and holder. Mains on-off switch. Plastic case and panel approx. $6 \frac{1^{\prime \prime}}{2} \times 3 \frac{3}{1 "}^{\prime \prime} \times 2^{\prime \prime}$.


View of the PCB, mounted directly on the meter terminals. The mains transformer and smoothing capacitor C3 are mounted in the box.

Meter M1 may be omitted and the terminations linked if an audio type alarm only is required. The audio output is surprisingly effective for general alarm purposes and indeed quite startling!
The prototype was proved effective with many inflammable gases including carbon monoxide, petroleum vapour, carbon 'tet', flux vapours, alcohol and many types of smoke all triggered the alarm.

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# Circuits for AUDIOAMPLIFIERS 

## PART 2

## F. G.RAYER G3OGR

## HEAT SINKS

The power dissipated in a transistor is the product of the current flowing through it and the voltage across the junction. The emitter/collector voltage is lower than the supply voltage, by the extent of the voltage drop in collector and emitter circuits. In low level stages, dissípation will be considerably less than the safe limit for the transistor.
In qutput stages, and some driver stages, the power dissipated in the transistor becomes quite large. This power causes heating, which can damage the transistor. This damage is avoided either by limiting operation conditions so that dissipation is low enough for the particular transistor when used without a heat sink, or by providing the transistor with a heat sink. A heat sink is a metal clip, plate, chassis or something similar, either flat or finned, thermally in contact with the transistor. It carries away heat from the transistor which can then operate at higher ratings. The heat is lost by convection of air currents, radiation and conduction.

The thermal capacity of the sink, or its ability to carry away heat, is arranged to suit conditions. The heat sink is an essential item and should not be omitted when specified. The chassis or metal case is often used, to avoid a separate heat sink.

It is sometimes necessary to provide electrical insulation between the transistor and its heat sink and a thin mica washer can be used. Sink surfaces should be flat for good thermal contact, with no rough holes
or edges which may penetrate the insulation. To obtain a low thermal resistance (good transfer of heat) silicone grease, available for this purpose, may be smeared on the surfaces.

## PNP AMPLIFIER

Fig. 13 is a directly coupled circuit for 14 V , either from an accumulator (nominally 12V) or derived from the mains. Pre-amplifier and driver stages are included. VR1 is pre-set and adjusted so that a meter placed in one supply lead reads about 900 mA . The output transformer is centre-tapped. With output stages drawing a heavy current, significant voltage is dropped and efficiency reduced if the transformer winding has an unnecessarily high resistance, so a large component intended for such circuits is necessary. Tr3 requires a heat sink which can be $5 \times 4$ in 16SWG aluminium, or a larger area obtained by using the case or chassis.

## THERMAL RUNAWAY

If operating conditions of a stage, or directlycoupled stages, are not sufficiently stable, changes in working conditions arise. The amplifier may shift into operating conditions which reduce gain, cause distortion, or increased collector current. With low power stages, emitter and collector resistors assist stabilisation, and limit currents to safe levels. But with higher powered stages, adequate safeguards in this way are not possible.


A directly coupled amptifier using an AD149 transistor in the output stage. The supply should be around 12 to 14 V . a car battery being an ideal source, the quiescent current of the whole amplifer being about 900 mA .





## A practical circuit for an audio

 amplifier having a push-pull output stage. Quiescent current is low so a gV battery is salisfactory as a power supply.In these circumstances, when external circuit resistance and operating conditions do not limit current to a safe level, thermal runaway may develop. Power dissipated in the transistor produces heat, and as the junction temperature rises the collector current increases even more. This, in turn, causes a further rise in dissipation and temperature which again increases current. This situation continues until the current and thus the junction temperature is limited by external circuit values or until the transistor is destroyed.

## THERMISTORS

Component R2 in Fig. 12 (Part 1) is a thermistor and its resistance falls as its temperature rises. The rise of temperature of R2 may be due to external conditions, to heat in the equipment or to the current passing through it, or both. An increase in ambient temperature or in operating voltage will cause the collector current of the stage to increase. But the effect on R2 will be to reduce the base voltage since the value of R2 falls. The drop in baseemitter voltage reduces collector current. The stage can thus operate satisfactorily over a wider range of temperatures and voltages than it could if the thermistor were not present.

## PUSH-PULL OUTPUT

If the audio cycle is divided into two, each peak can be taken to an individual transistor and can drive it into conduction. The outputs of the transistors can be combined to obtain the original audio signal. The advantage of this method lies in the fact that both transistors can have a low resting or quiesoent current.

Each transistor of the output pair draws current in turn so this is called a push-pull stage. As the transistors are each dealing with only about one-half of the audio cycle, they are operating in Class B. (This distinguishes operation from Class $\mathrm{A}_{4}$ where a single transistor deals with the whole audio cycle.)

Many circuits with Class B push-pull output dispense with transformers, but Fig. 14 is a useful circuit, representative of many transformer coupled
amplifiers. Its push-pull operation is also easy to understand, while it has the advantage that the DC operating conditions of each stage are isolated from other stages. The output is around $500 \mathrm{~mW}\left({ }_{2} \mathrm{~W}\right)$.

Tr 1 is a Class A pre-amplifier, and can be fed from a pick-up or tuner. R2 and R3 set base conditions, which R5 is the emitter bias resistor. The audio signal is developed across the collector load R4, and is coupled to $\operatorname{Tr} 2$ by $\mathrm{C} 3 . \operatorname{Tr} 2$ is also a Class A amplifier, and termed the driver as its purpose is to drive the output pair, $\operatorname{Tr} 3$ and $\operatorname{Tr} 4$. The primary of the driver transformer Tl furnishes the collector load for Tr2.
The primary of T 2 is centre-tapped. Bias conditions of $\operatorname{Tr} 3$ and $\operatorname{Tr} 4$ (from R10 and R11) are such that Tr 3 and Tr 4 collector currents are small (say, 10 mA for the pair). Since T1 secondary is centre-tapped, $\operatorname{Tr} 3$ and $\operatorname{Tr} 4$ are alternately driven negative into conduction. Outputs are combined by the centretapped transformer T2.

Current drawn by the output pair depends very much on the audio signal level, rising to peaks of 30 to 60 mA or more, according to volume. (This can be observed by placing a meter in one battery lead.) The average current is, however, much lower. Battery life is lengthened, and the two transistors can deal with appreciably more power.

R13 provides negative feedback (see later) over $\operatorname{Tr} 2, \mathrm{~T} 1$ and $\operatorname{Tr} 3 / 4 . \mathrm{C} 7$ and R14 are to reduce the effects of phase shifts at higher frequencies. Connecting R13 should cause a drop in volume but an improvement in quality. If this effect is reversed or oscillation arises, take R13 to Tr4 collector instead of to Tr'3. With the values shown for the " $1_{2}$-watt" amplifier, a maximum output of 500 mW is obtainable, and no heat sinks are necessary.

A wide range of transistors can be used in this circuit and the operation of each stage can be checked and adjusted if needed. R10 may be a preset resistor and adjusted so that $\operatorname{Tr} 3 / 4$ draw about 10 mA , with no signal. If the value of R10 is too high, current will be low and cross-over distortion caused. On the other hand, if R10 is low in value, the quiescent current is high.

Fig. 15 is typical of circuits using a driver transformer with a split secondary. Each output transistor
operates on approximately one half the supply yoltage. The circuit has the advantage that no output or speaker transformer is necessary. Trl is the driver stage, with transformer Tl as described ealier. R6 and R7 provide base bias for Tr2, while R8 and R9 provide base bias for $\operatorname{Tr} 3$. Each output transistor has its own emitter resistor R10 and R11.
It is essential that the halves of the transformer secondary, Sec. 1 and Sec.2, are correctly phased for the transistors to operate in push-pull. If not, connections to one secondary must be reversed. The negative feedback network C5/R12 is to improve response, and operates over the whole amplifier.
The circuit allows an identical or matched pair of transistors to be used, as for Fig. 14. Maximum output is reduced but battery drain is smaller than that for Fig. 14. Overall gain is sufficient if a tuner is used as input. For other purposes, any of the lowlevel pre-amplifier circuits with a positive earth line can be placed before Trl.

duction. As phase shifts may arise near upper frequency limits, it may be necessary to suppress these, as by using C7 and R14, Fig. 14, so that they are not applied to an earlier stage in positive phase. Cir cuits with no transformers are arranged so that the feedback is correct.


## NEGATIVE FEEDBACK

A negative feedback circuit of some kind will be seen in virtually all but the very simplest amplifiers. This circuit takes a part of the amplifier output and applies it to the amplifier input. The feedback must be arranged so that it is out of phase. If it were fed back with the same phase as the original input signal, it would cause uncontrollable oscillation.

An amplifier does not normally give exactly the same amplification at all frequencies, because components in it (especially capacitors, transistors and transformers) do not have the same characteristics over a wide range of frequencies. Thus, some frequencies may be emphasised. If so, they are of greater amplitude in that part of the signal fed back to the input, and so help maintain a more level response at all frequencies.

Negative feedback may be over a single stage or may extend over two or more stages. In some circuits, negative feedback is purposely made frequency sensitive. This is one method of obtaining tone control, or in securing more satisfactory repro-

## COMPLEMENTARY SYMMETRICAL OUTPUT

With a PNP transistor, a negative-going base input increases the collector current. The NPN transistor, however, requires a positive-going base input to increase collector current. So by combining PNP and NPN transistors it is possible to make a pushpull output stage which does not require a centretapped driver transformer or other means of phase splitting. Such circuits are used extensively.

In the basic circuit, Fig. 16, forward bias is developed across R2 in conjunction with R1 and R3. As explained, the audio input will cause first one transistor to conduct, then the other, for successive half cycles. The speaker is connected to the emitters, and two batteries (or a tapped battery) provides a supply for the transistors.

In Part 3 next month the matching of speakers to amplifiers is dealt with together with the use of IC's in audio amplifiers.

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by Eric Dowdeswell G4AR

NIGEL Roberts GBJEF of Prescot got a bit hot under the collar over my comments on repeaters so here is an agreed version of his comments.
"I feel that I must take strong exception to Eric Dowdeswell's unwarranted attack on repeater users. (On the Air, Amateur Bands, January). I am not an FM or repeater nut, my main interest lies with SSB DX on 144 MHz but I do use it for local working and the repeaters for mobile. There is abuse, mostly 'things' without a callsign, making funny comments which admittedly don't reflect much credit on FM and repeaters. But the public service potential, especially with Raynet (Emergency Network) does. It is not my intention to perform a defence of repeater operation, the fact that a large number of amateurs use them justifies their existence but to but to point out that there are abuses on other bands and methods of operation. Exoessive power is one. On 160 m where the limit is 10 W it has been reported that one station has a capability of 1 kW ! Bad manners is another.
If the amateur movement starts publicly squabbling over the best way to operate then in 1979 we might turn round and find no repeaters, and no other bands either."

Now I must point out to Nigel that I had been asked by several readers for my views on repeaters, and I gave them, but I did not even mention the users! I could have said that the service is often abused but I didn't. You did, Nigel! (The problem with abuse, whatever band is involved, is that the Home Office must catch the effender in the act, and that is patently just about impossible!) The number of amateur licencees is around 23000 but how many use the repeaters for their intended purpose? A few hundred, at the most. And how many of these are members of the RSGB? So how does this 'large number' justify all the trouble that the RSGB has gone to to get the repeater service established? They should distribute their services to suit the majority not the minority!

Several readers have commented on the continuing improvement of the 15 m band. Steve Cottis A8961 (Harrogate) wonders why so many stations stay on 20 m fighting the QRM when they could work just as much DX on 15 m ! So it's back to my long-held theory that the bands are never as bad as we make them out to be but just lack of activity. Steve says that 15 m has opened regularly to west and south Africa on most week-ends but, of course, it is probably like that every day but, unfortunately, most of us have to earn a living! Steve is very pleased
with some of his QSL returns lately which included M1D, D4CBS, A6XR, 5U7AG and F08DO.
J. Taylor in Cheadle Hulme, Cheshire, bemoans the fact that, in his opinion, there is no good textbook for the RAE. He'd like one written in the style of the ARRL Handbook. He goes on to make some blush-making comments on this column which have been duly framed! He'd also like to see some good articles in PW on general coverage communication receivers as 'there would appear to be a gap between the crystal set and the Video-Writer! Point taken! Paul Bradbeer writing from Braunton in N. Devon was pleased to find that J. R. Beer (PW Feb) lives only five miles away. Have a look at your names OM, you could be related! Anyway, Paul wonders if there is enough activity among types like himself to form a club in that part of the world. He hopes to be getting a PR40 preselector for his CR70A very soon but at the moment his life revolves around mock ' O ' levels!

Paul Barker BRS34898 DXing from Sunderland added TG9 on 20 m for a new one and KP4 and FP8 on 80 m to swell the list there. However Paul is now busy playing with his new baby, a Robot 70C SSTV monitor, for which is seems no praise is too great. Apparently it is regarded as the Rolls Royce of SSTV. His previous monitor is now up for sale so ring him on Sunderland 226883 or write to 11 Dipton Gardens, Tunstall Estate, Sunderland, if you have some cash to spare. Down in the Channel Islands Alan Troy sports a Trio 9R59DE with an indoor aerial in a downstairs flat so he hasn't exactly got the best start for DXing. He's been fiddling around for about ten years and now feels he really ought to have a go at getting his ticket. What better OM, if you don't want to waste time listening for the unobtainable DX! For those who may be out of touch Alan points out that the general C.I. prefix of GC has now become GJ for Jersey where Alan resides and GU for neighbouring Guernsey. The Jersey ARS station GJ3DVC has had massive pile-ups, doling out the new prefix!

Calling you out there in Leeds land! Bob Firth G3WWF Hon. Sec. of the Wakefield and District RS tells me that they have recently conducted a wide-ranging survey, in the form of a questionnaire, into the likes and dislikes of members and potential members, with a view to rejuvenating the club. A special all-comers meeting has already been held but if you slipped through the net and live in the district why not contact Bob on Leeds 825519. Meetings are held at 1930 at the Youth Centre, Ings Road School, Wakefield. On 15 March it's John Hey G3TDZ speaking and showing off his constructional projects, on 29 March a ragchew and operating night while 12 April should be quite popular, it's a Junk Sale.

If your QTH is in the Derby area then you should be off to the Derby and District ARS AGM at 1930 on 16 March to be held at the clubroom at 119 Green Lane, Derby. Old news now, but at their Annual Dinner in February the new President of the RSGB, Lord Wallace of Coslany was their guest of honour.

John Wood G3YQC of the British Amateur Television Club has just sent me details of their new publication A Guide to Amateur Television. Written by members of BATC it covers all aspects of receiving and transmitting amateur TV signals including colour TV and closed circuit systems. The price for 112 pages of concentrated information is
just $£ 1 \cdot 25$ post paid to members, which strikes me as remarkable value today! For non-members it's $£ 1 \cdot 75$ so it might pay one to join the BATC at the same time! Orders to BATC Publications, 64 Showell Lane, Penn, Wolverhampton, Staffs WV4 4TT.

In Leyland, Lancs, Paul Cowburn has had continuing trouble with his CR300/2 receiver but a multimeter and soldering iron for Christmas helped him to put it right and he is active on the bands again. His efforts to learn the code using records are bearing fruit as Paul has been copying a lot of CW stuff from 160 m to 20 m . Paul was one of those fooled by GJ3DVC of Jersey, mentioned earlier! Eric Hill of Guildford got bitten by the bug and returned to the bands about a year ago with a National Panasonic RF1700 but soon graduated to a Hammarlund HQ100A plus a Codar PR30, which together with about l70ft of wire, seems a pretty good set-up. First DX for the New Year was ZD8EW on 80 m SSB so let's hope that is a portent of good DX to come!

John Reynolds in Bridgend very wisely decided to stay on the amateur bands having abandoned the broadcast frequencies. He has completed a regenerative set which can copy SSB so we can look forward to some logs soon. A letter from Neil Braeman, near Horsham, sent me details of the December RAE paper. Unfortunately, I can't tell you the pass mark Neil so you will have to wait out the couple of months or so. Let's all hope you made it Neil since you have now passed the morse test! Congrats! D. Peck BRS37621 of Cambridge sends in a detailed log of stuff copied in a new mode for this column, RTTY! He's logged over 30 countries so far but doesn't mention the gear. So let's have some more info OM, please.

Now lads and lassies, there were several logs received after the deadline this month. Please make it a bit earlier or, regrettably, they will not make the column.

## Log Extracts

P. Barker:- 20 m TG9DV TG9QK 5V7WT 80m FP8DX KP4AST SSTV 20m DJ8LV EA2JO F3RT HA5KBM K4TGC SP3PJ VE3PT W8KZM YUINWJ

A Troy:- 80m CN8BF CT3BM D4CBS (Cape Verdes Is.) EA8CR EA9FE (Mellila) FP8ZZ HK4DF KZ5HP PZ1AC W6NLZ 9G1JX
B. Harrison:- 80m FG7AO VP9CP 7Z8VA (Saudi Arabia) 9M2MK 20m FY7AW P29GR XT2AE
S. Cottis:- 80m EA8CR EL7F KP4AST KZ5HP 20 m YB1KW 9J2PC 15m CT2BB EA8CR EA9FC
P. Bradbeer:- 80m EA8JP JA6DG 20m YN1CCA ZL3FV AH6UM (KH6)
P. Cowburn:-160m CW K1PBW W3HXK W4BRB W8LRL KV4FZ 80m FP8ZZ HI8LC HK4DF JH0BQU JX2FL KZ5HP PY7BZD UK9AAN W7KW 5Z4GX 9Y4NP 40 m PY2CAR PY8AKA LU4MEE VK3ZL ZL1AIZ 6W8AAD 20m KG6SW VK5BC YB4ACJ ZLIGG
D. Peck:- 15m TU2GG WB2CJS/HZ1 VE2ZN/SU WB6EWH/VQ9 DK5EC/ET3 WB4BWG/KG6 20m RTTY AC3KEK CT1EQ EA8IT VE2QO VK2SG WB6EWP/VQ9 3A2FB 9H1EL 9M2MW
N. Braeman:- 80 m CT2BS CT3BM KZ5HB VP2LDU XE1KB 20m WB6EWH/VQ9 8P6AH
E. Hill:- 80 m FM7WS JY3ZH PJ2FR VP2LCT VP2SN ZD8EW 5B4PW 9K2DR

All stations are SSB except CW, in bold, and RTTY in italic.


## SHORT WAVE BROADCASTS

by Derek Bell

IRECENTLY stated with much conviction that Radio Australia could not be heard around midday in the UK. However, the unprecedented conditions over Christmas and New Year period proved me wrong. This is pointed out by Alan Spencer from Ryhall, in Lincolnshire, who, on his Heathkit SW717 plus a long wire, pulled the signal in on 9560 from 1230 to 1300 , admittedly at low strength but there, nevertheless. This disturbance was one of the strongest winter upsets that I have experienced and it lasted for around six days, from Christmas Eve onwards.
This month's overseas letter comes from Radio Finland who sent me their schedule from March to May and they offer a service from 0930 to 1000 on 9550,11755 and 15270. From 1430 to 1500 on 6120 , 11755 and 15110 and from 1900 to 1930 on 9720 and 11755. From 2000 to 2030 it's 6120, 9550. It seems that Radio Finland has thrown in the towel as far as North America is concerned. They admit that the Pori transmitter is too far north to give an adequate signal to the US and Canada, due to the beam having to cross the auroral zone which attenuates the lower frequencies.

Bournemouth School has among its after-hours activities a thriving radio club and the boys sent along a very wide ranging $\log$ pulled in on their Eddystone 730 and AR88. They managed to gather the following:-

| Radio Pekin | on 7590 at 1700 |
| :--- | :--- |
| Radio Cairo | on 17920 at 1436 |
| WYFW | on 15110 at 1745 |
| Radio Sweden | on 9630 at 2300 |
| R. Nac. Do. Brasil on 11780 at 2010 |  |
| Radio Athens | on 15345 at 1520 |

Mind you, if the lads were to admit the truth their master Mr Thomas must come in for some credit! Word is coming in of the assorted literature that the stations have been sending out recently. Jim Coombs of Walsall has had two copies of a questionaire from Radio Kiev pointing out that 1977 is the 60th anniversary of the great October revolution and what would you like to hear from Radio Kiev, all in one.sentence! I personally have had two similar style items, from Radio Poland and Radio Sofia.

David Wyatt of Oswestry has come up with a log of rarely reported items, and, although few, are stations that seldom seem to figure in reports these days.

> Radio Pyongyang on 9678 at 1605 Korean BS Seoul on 9640 at 1140 UNRS

The last service is the broadcast of the proceedings of the United Nations, relayed through Voice of America and of course it is only available during sittings of the General Assembly of the U.N. A Vega Selena hung on the end of a 78 ft long wire is the equipment that young David Birch uses. He


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has written giving the background to the Radio Athens English service, David logged them on 17830 at 1230 and has forwanded the address which is:Director of Technical Services, EIRT, 16 Mourouzi Street, Athens, Greece. The latest schedule is as follows:-

```
0115 on 11730
0230 on 9760 and 9750
1230 on 17830 and 15345
1530 on 17830, 15345 and 11730
1930 on 9675, 7215 and 6140
0930 on 15345 and 9530
```

Since starting this month's column I have had the pleasure of being able to assist, in a practical way, a member of the DXing fraternity. Some time ago Jim Coombs asked for any information on an R1392D receiver. By the good offices of another reader a circuit diagram arrived chez Bell, so Jim is happy. On that happy note I will close this month and wish 73 s to you and yours.


## MEDIUM WAVE DX

by Charles Molloy

VINTAGE Radio is not usually associated with DXing so readers may be surprised to learn of transatlantic reception this winter by a DXer using a vintage medium wave receiver. Tudor Rees Vintage Services, who advertise in PW, mention in the January issue of their Antique Wireless Newsheet, their reception the previous month of WINS New York City on 1010 kHz with a 1930 TRF receiver. Quite a remarkable achievement! Transatlantic medium wave reception was commonplace in the early days of radio, and wireless magazines of the mid-twenties made frequent reference to it. The circumstances were quite different then. Antifading transmitting aerials had not been invented and the flat-topped aerials of the day radiated lots of sky wave. There was little interference either electrical or from other stations and the unselective wireless sets in use could really pull in the DX. The writer is attempting to refurbish a receiver of circa 1924 vintage with the intention of trying it under the very different conditions that prevail today. Has any reader tried DXing with a vintage radio? The most likely channels for North America are 710 kHz , 930 kHz and 1010 kHz and at this time of year it will be 0100GMT before the path is open.

It appears that we have now reached the minimum of the current sunspot cycle, a period that favours DXing on the lower frequencies, and logs are now recording some of the less common North Americans. F. T. Shortridge of Llandudno Junction has his AR88 back in circulation and he reports hearing, when using a loop, CHNS in Halifax NS on 960 kHz , CKEC New Glasgow NS on 1320, WVOJ Jacksonville, Florida also on 1320, WEVD New York on 1330, WPOP Hartford on 1410 and WOKO Albany NY on 1460. G. Braithwaite who writes from Belfast
used a Drake SW4A and loop to log VOCM St John's, Newfoundland on 590 kHz , WHAS in Louisville, Kentucky on 840, WHDH in Boston on 850 , CFRB Toronto on 1010, WWWE Cleveland, Ohio, on 1100 and WHAM in Rochester NY on 1180. Michael Tallent does his DXing in Stafford with a Realistic DX160 and a 300 ft long wire. His log includes KDKA Pittsburg on 1020 and he mentions that he can pick up many distant broadcasts which he cannot identify. All North American broadcasters on the medium waves are allotted call letters which they are obliged to give over the air but with the characteristic slow fading that occurs on this band it requires time and patience on many occasions in order to obtain a positive identification.

Reader James Birkett of South Kensington sends a report of stations heard between 1245 and 1355GMT during December using a Telefunken portable receiver attached to his TV aerial! Broadcasts from as far away as Monte Carlo on 1466 and Andorra on 701 were heard in daylight. In mid-winter, when the sun rises only to a low angle above the horizon, there are occasions when the D layer in the ionosphere only partly reforms at sunrise. Normally the D layer reforms completely then and absorbs signals in the medium wave band, inhibiting long distance reception. At the time of the last sunspot minimum in 1965-66 North American DX was heard as late as 1000 because of this phenomena but it is very unusual. Readers might follow James's example and try their TV aerial for MW DXing as it makes a good vertical for medium wave reception. Try connecting the receiver to the inner and the screen of the co-ax lead in turn for optimum reception, or both!

Monday morning is a familiar time for DXers in North America. Many high power outlets in that area, though nominally on the air 24 hours a day, close down for transmitter maintenance between midnight on a Saturday and 0500 on Monday, local time. DXers in the United States regard Monday morning as the best time of the week for local and foreign DXing. In the UK too, advantage can be taken of this period of quiet, provided that the DXer is prepared to be out of bed at five in the morning! Owing to the different time zones in operation in the USA, midnight in New York corresponds to 2300 in Chicago and 2100 in Seattle and 0500 in the UK, which means that stations on the east coast close down earlier than those further inland. When conditions are good it is worth trying the band at 0500 on a Monday. KING in Seattle on 1090 has been heard on a Monday with WBAL in Baltimore off the air, KEX in Portland, Oregon on 1190 was logged by the writer with WOWO in Fort Wayne, Indiana off and WOKJ Jackson, Mississippi on 1550 with CBE Windsor, Ontario off the air.


## by Ron Ham

ON December 29th, Cmdr. Henry Hatfield (Sevenoaks) located two sunspots and an angry-looking filament with his spectrohelio-
scope (see Amateur Photographer, March 24, 1976) and, true to form, both he and your scribe recorded strong radio noise at 136 MHz from this event the following day. Henry found another pair of sunspots on January 11th and, like the earlier ones, these were active and emitting radio noise at metre wavelengths. Although several tiny bursts of solar noise were recorded between the 12th and 17th the strongest and most continuous was heard on the 12 th at both 95 and 136 MHz , and again on the 14 th , but this was limited to the higher radio frequency. (Reports of exceptional HF DX during this period would be most welcome.)

From the heavens came the Quadrantid meteor shower (Jan 1 to 6), and this looked good by radio. As the earth approached the shower on January 2nd the writer's equipment counted 5,515 "pings", of signal from the Gdansk broadcast station ( $70 \cdot 31 \mathrm{MHz}$ ) during the 15 -hour observation period ( 0800 to 2300 hr daily). On the 3rd, peak day, more than 28,000 "pings" were recorded which should have given excellent conditions for the meteor scatter enthusiasts. As usual the daily rate of "pings" reduced pro rata as the earth passed through the main swarm of particles; 4 th was $11,100,5$ th was 8,612 and 6 th was 6,858.

Throughout the peak day "pings" of signals were also heard simultaneously with Gdansk from a Russian TV transmitter on $49 \cdot 75 \mathrm{MHz}$ (R1) which shows that signals can bounce off meteor trails at a variety of radio frequencies. Last month's report about the Geminid meteors prompted Chris Webster G3TBJ (Ringwood) to write about meteor "pings" on signals in Band II, During the early hours of the morning Chris often turns his 6-element beam (with 3N200 pre-amp) north-east for a listen to Capitol Radio on $95 \cdot 8 \mathrm{MHz}$. "This provides a splendid stereo signal", says Chris, "except for interruptions every few minutes by chunks of German speech or music", probably from the GDR 50 kW DDR1 from Berlin. Chris suggests that readers without 4 m gear wishing to observe radio meteor "pings" should tune to a quiet spot in Band II and listen for a German station. Chris leaves his receiver tuned around 96.5 MHz for this purpose. The next meteor shower to watch out for is the April Lyrids, April 19 to 24, peaking 22nd.

Whilst on the subject of Band II, another interesting letter came from Nick Taylor (Sunbury-onThames) who has now added a barometer to his station equipment and will be watching for those atmospheric pressure changes which bring the VHF DX. He already has a good score on Band II judging from the copy of his log which includes more than 20 each German and French broadcast stations, plus a dozen Italians; and a few Dutch and Spanish, all between 88 and 104 MHz , and of that lot 32 broadcasts were heard in stereo! A most interesting list Nick and one that we can refer to when your future reports come in. Needless to say there is a good selection of both G and GW DX in his log. You are dead right, Nick, the VHFs were quiet in January, however there was a pressure change between the 6th and 9 th which gave a mild "lift" to parts of the VHF spectrum.

During the evening of the 6th, Alan Baker G8LGQ (Newhaven) was listening to the Kent repeater, GB3KR Dover, and heard a station in Ipswich say that he was receiving a colossal signal from the repeater as well as signals on "two" from both PA0 and DL. It is interesting to note that there has to be a slight lift in conditions before G8LGQ can use
the Kent repeater. At midday on the 8 th, the writer received a good picture from the IBA transmitter at Lichfield on Channel 8 ( 189 MHz ), but this did not last long.

Alan Baker is a member of the Brighton Repeater Group and reports that the equipment for GB3BR on 70 cm had satisfactory tests on January 15th and 16th and by now should be on the air permanently (RB6, input $434 \cdot 750$, output $433 \cdot 150 \mathrm{MHz}$ ).

Although the recognised sporadic-E "season" in Europe is between mid-April and mid-August, our fellow enthusiasts "down under" had sporadic-E disturbances throughout the Christmas period. A most welcome letter, dated January 8th, came from Anthony Mann, a reader in Applecross, Australia, who says "we are at present enjoying an active sporadic-E season which has seen Malaysian TV (Channels E2 and E3) received here and Samoan TV (Channel A2) received in Sydney on Christmas day. The distances involved being 2,400, 2,600 and 2,700 miles respectively."
Anthony is a keen TV and VHF DXer and regularly monitors the 10 m band looking for any unusual activity which might indicate that sporadic-E or high MUF-F2 is present. The writer is always pleased to hear about VHF activity from any part of the world, because, apart from the interest for our readers, it all helps to build up a general pattern which aids the study of propagation. The only International Beacon Project station that Anthony has heard so far is $3 \mathrm{~B} 8 \mathrm{MS}(28 \cdot 190 \mathrm{MHz}$ ) in Mauritius and he asks for the frequency of the others on "ten": DLOIGI is on $28 \cdot 195 \mathrm{MHz}$, GB3SX on $28 \cdot 185 \mathrm{MHz}$ and 5 B 4 CY on $28 \cdot 180 \mathrm{MHz}$.
Your scribe heard a 549 signal from the Cyprus beacon (5B4CY) at 0826 on January lst and again at 0929 on the 7 th and would like to hear more from readers about their reception of the 10 m beacon signals.

It is worth remembering that there is a great deal of enjoyment to be had and a lot of knowledge about propagation to be gained by listening in, or taking part in the RSGB VHF/UHF contests and there are some good ones in store; 144 MHz Open on March $5 / 6$, 432 MHz Open on- March 20, 70 MHz Open in April $2 / 3$. So what about it readers, have a go, and good luck!

Thanks again for your letters, several of you have said that you are pleased to see a VHF report again in PW so keep on writing and between us we will make it an interesting feature for both our regular readers and for science in general.

## BROADCAST BANDS

Short Wave Reports by the 15th of the month to Derek Bell, c/o Practical Wireless, Fleetway House, Farringdon Street, London, EC4A 4AD. Medium Wave Logs to Charles Molloy, 132 Segars Lane, Southport, PR8 3JG.

## AMATEUR BANDS

Logs covering any amateur band/s in band/ alphabetical order by the 25th of the month to Eric Dowdeswell G4AR, Silver Firs, Leatherhead Road, Ashtead, Surrey, KT21 2TW.
VHF
Reports on VHF matters to Ron Ham, Faraday, Greyfriars, Storrington, Sussex RH20 4HE.


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10 amp type operated by the trips，thas 15 10 amp type operated by the trips，thas 15 circulta may be changed per revolution．
Drive motor in mains operated 5 revs per
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soldering frons unasable or very slow $\ln$ operation and also upsets the working of other Instruments Fiven In normal weather if you put a long line thrmoh tarndist wht werkkhop you mng finf that $1 \mathrm{~h} . \mathrm{colpage}$ is luw．lue to valtilec ilrop nlong the lito wot that compenmation tranaformpr may be the
（＇hpaj）way of regtoring voleage to normal．The chpai）way if restoring voltage to normal．The other ulternative is to install a much heavier cable．
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Cormplete with P220 Powner Supply LISTgis：50

## DIY SPEAKER KITS

EASY－TO－BUILD WITH ENCLOSURE
Specially designed by RT－VC for cost－conscious hi－-1 enthusiasts，these kits incorporate two teak simulate enclosures two EMI $13^{\prime \prime} \times 8^{\prime \prime}$
（approx．）woofers，two
tweeters and a pair of matching cross－ overs．Easily constructed，using a few basic tools．Supplied complete with an easy－10－ follow circuit diagram，and crossover components．Input 15 watts rms． 30 watts peak，each unit 22550 Cabinet size $20^{\prime \prime} \times 11^{\prime \prime} \times 9^{\prime \prime \prime} 2^{\prime \prime}$

PER PAIR （approx）

D \＆p $£ 550$
15－WATT KIT IN £ $17.00{ }^{\text {PER }}$ STERED CHASSIS FORM ${ }_{\text {f3 } 40 \mathrm{P} \text { \＆P PAIR }}$ When you are looking for a good speaker，why not build your own from this kit．It＇s the unit which we supply with the above enclosures．Size $13^{\prime \prime} \times 8^{\prime \prime}$ （approx．）wooter ，，EMII＇weeter． and matching crossover．Powe handining capacity 15 watts rms 30 watts peak

## COMPACT＇FOR TOP VALUE

How about this for incredible bookshelf value from RT－VC！A pair of high efficiency units for only $£ 7.50$－just what you need for low－power amplifiers．These infinite baffle enclosures come to you ready mitred and professionally finished Each cabinet measures $12^{\prime \prime} \times 9^{\prime \prime} \times 5^{\prime \prime}$（approx deep，and is in wood simulate Complete with two $8^{\prime \prime}$ （approx．）speakers for max．$£ 750$ power handling of 7 watts．－p\＆p $£ 170$


Superb Viscount IV unit in teak－finished cabinet．Silver fascia with alumınum rotary controls and pushbuttons．red mains indicator and stereo jack socket．Function switch for mic，magnetic and crystal pick－ups，tape，tuner and auxiliary．Rear panel features two mains outlets，DIN speaker and input sockets，plus fuse． $20+20$ watts rms， $40+40$ watts Deak．


SYSTEM 1B For only 880 ，you get the $20+20$ watt Viscount IV amplifier a pair of our 12－watt－rms Duo Type llb matched speakers，a BSR MP 60 type deck complese with magnetic cartidge．de luxe plinth $\quad \mathbf{8 0} 0^{00}$ and cover

SPEAKERS Two models reak veneer． 12 watts rms， 24 watts peak． $181 / 2^{\prime \prime} \times 13 / h^{\prime \prime} \times 74^{\prime \prime}$ approx £34 ${ }^{\text {PER PaiR }}$ Duo ili， 20 watts ms， 40 watts peak $27^{\prime \prime} \times 13^{\prime} \times 11^{\prime \prime \prime}$ －approx ${ }^{5} 52$


## 4． $4 \times 4$ STEREO AMP

 ，w KIT f14．50 KIT f14．50 p\＆pe200 ＊For the experienced constructor who wants to design his own stereo，kit includes all necessary components including constructors manual Plus Parr of easy to build 4 watt speakers in kit form with teak simulate finish cabinets $12^{\prime \prime} \times 9^{\prime \prime} \times 5^{\prime \prime}$ approx
## 痽

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Complete with speaker，baffle and fixing strip．The Tourist IV tor the experienced constructor only The Tounist IV has five push buttons，four medium band and one for long wave band．The tuning scale is illuminated and attractive small aluminium control knobs are used for manual tuning and volume control． The modern style fascla mas been designed to blend with most car interiors and the finismed radio will slot into a standard car radio aperture Size approx7．2．2．4．Power Supply Nominal 12 volts positve or negative eart nternallyi Power Ouptut 4 watts into 4 onms

SYSTEM 2 Comprising our $20+20$ watt Viscount IV amplifier：a pair of our large Duo Type Ill matching speakers which handle 20 watts rms each，and a BSR MP 60 type deck with magnetic cartridge，${ }^{〔} 9 \mathbf{2 0}^{00}$ de luxe plinth and cover

92

## TURNTABLE Popular BSR

 MP 60 type，complete with magnetic cartridge，diamond stylus，and de luxe plinth and cover． $29^{00}$


## 35－WATT DISCO AMP

Here＇s the mono unit you need to start off with．Gives you a good solid 35 watts rms， 70 watts peak output．Big features include two disc inputs，both for ceramic cartridges，tape input and microphone input．Level mixing controls fitted with integral push－pull switches Independent bass and treble $\mathbf{\varepsilon} 750$ controls and master volume． $2 \int_{-p \& p: 150}$


PORTABLE DISCO CONSOLE
with built－in pre－amplifiers
Here＇s the big－value portable disco console from RT－VC！It teatures a pair of BSR MP 60 type auto－return，single－play protessional series record decks．Plus all the controls and features you need to give fabulous disco performances．Simply connects into your existing ${ }^{〔} 64{ }^{00}$ slave or external amplifier．
＋p\＆p§650


70 \＆ 100 WATT DISCO AMPS

Brilliantly styled for easy disco performance！ Sloping fascia，so that you can use the controls without fuss or bother．Brushed aluminium fascia and rotary controls．Five smooth－acting，vertically mounted slide controls －master volume，tape level，mic level，deck level，PLUS INTER－DECK FADER for perfect graduated change from record deck No． 1 to No．2，or vice versa．Pre－fade level control （PFL）lets YOU hear next disc 170 WATT | before fading it in．VU meter |
| :--- | :--- |
| monitors output level 70 watts | $\mathbb{Q 4 0 0}^{\mathbf{4}}$ watts rms 140 watts peak output

All the big features as on the 70 －watt disco amplifier，but with a massive 100 watts rms 200 watts peak output power．

 PLAYER MECHANISM Requires some atten－
tion．Complete with built $\mathbf{E}^{95}$ in pre－amp，A．C． $240 \mathrm{~V}+\mathrm{p} \& \mathrm{p} £ 150$

## ALL PRICES INC．VAT．AT $12 \frac{1}{2} \%$

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[^2]:    Send $75 p+10 p$ (post \& packing) to: Chief Cashier (PW Tele-Tennis), IPC Magazines, King's Reach Tower, Stamford St., London, SE1 9LS.

[^3]:    

[^4]:    Full size reproduction of the LED board, showing the foil side above with the component overlay below. A certain amount of pressure is exerted on the LED's when the lid is offered up to them, this tends to lift the track. It is recommended that a stiff piece of card be inserted (between the leads)

[^5]:    version ET.50.

[^6]:    LOW-COST I.C. MOUNTING for any size DIL package. 100 Soldercon sockets 65 package. 7 and 8 hole plastic sockets 65 ,
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