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## BACK NUMBERS

We regret that we are unable to supply back numbers of Practical Wireless. Readers are recommended to enquire at a public library to see coples. Requests for specific back numbers of Practical Wireless and Television only can be published in our CQ Column.
NEWS \& COMMENT
216 FORMATIVE HABITS-Leader article and August preview
217 NEWS . . NEWS

$\qquad$ ..... NEWS
229 TELEVISION-This month's attractions
230 PRODUCTION LINES-Products reviewed by Colin Riches
234 P PRACTICALLY WIRELESS-Light hearted view by Henry
239 NEXT MONTH in Practical Wireless
254 ON THE AIR-Reports on Broadcast Bands
254 Short Wave-Malcolm Connah
257
257
258 Amateur Bands, Short Wave/V.H.F.-David Gibson G3JDG
261 DEAR SIR-Readers' letters to the Editor
262 HOTLINES on recent developments-Ginsberg
CONSTRUCTIONAL
218 EXTRA BANDS F.M. RECEIVER-F. G. Rayer
235 I.C. PROCESS TIMER (0-99 seconds)-G. H. Woods240 TREMOLO GENERATOR for guitars and organs-R. Morgan246 3-BAND TRANSMITTER, Part 2-V.F.O.-F. G. Rayer G3OGR260 TAKE 20 No. 50 Electronic Siren-David Andrews265 "P.W. TRICOLOUR" Mood Lighting System, Part 4-J. Budek
OTHER FEATURES225 MAGNETIC AMPLIFIERS, Part 1-I. R. Sinclair233 I.C. OF THE MONTH, TBA435 Voltage Regulator-

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250 GOING BACK Equipment of yesteryear-
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Individual Cianged Controls: Bass, Treble, Volume and Balance. Primed circuit construction employing 10 Transistors plus Diodes. Output rating I.H.F. M. Frequency range $20-20,000 \mathrm{c} . \mathrm{p} . \mathrm{s}$. Bass Control th 12 db . Treble Control $\pm 13 \mathrm{db}$. Selector switch for P.U. or Tape/Radio. For loudspeaker output impedances of 3 10 15 ohms. For standard $200-250 \mathrm{v}$. A.C. mains operation. Attractive Black and Silver finished metal tascia plate and matching control knobs. COMPLETE KAT OF PARTS INCLODING
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8 ohma, diven smooth realintic sound output 8 ohme. Aven smooth reatintic sound out put

Carr. 30p

## HI-FI SPEAKER ENCLOSURES MODERN DESIGN

 Teak veneer finish. Acounticaliy linech. Sizes approx. Cars. 3opp. pereJE8 Size $16^{\prime \prime} \times 14^{\prime \prime} \times 9^{\prime \prime}$. SE8 For optimum perform Pressurised. Gives pleasing results with any
8 in . Hi-Fispeaker. $\leq 5.50$ ance with any 8 in .
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For Mullard 510 , 6.3 v . 4 a .
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and Cassette Tape $28 p e x t r a$ if required．


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[^2]
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## 4channel decoder modules



Specifications

| Input Impedance | 40 KOhms |  |
| :--- | :--- | :---: |
| Output Impedance | $<300 \mathrm{Ohms}$ |  |
| Phase Shift Network $90 \pm 10^{\circ}$ | $20 \mathrm{~Hz}-18 \mathrm{KHz}$ |  |
| Nominal Gain | Unity |  |
| Frequency Response $\pm 1 \mathrm{~dB}$ | $5 \mathrm{~Hz}-100 \mathrm{KHz}$ |  |
| Nominal Input Level | 250 mV rms |  |
| Nominal Distortion | $0.025 \%$ |  |
| Clipping Point | 2.5 Vrms |  |
| Distortion before Clipping | $0.08 \% \mathrm{max}$. |  |
| Front Separation | $>60 \mathrm{~dB}$ |  |
| Rear Separation | $>20 \mathrm{~dB}$ |  |

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features: exclusive wide band phase shift network; distortion typically $0.04 \%$. Ideal for use with other kit projects and even with the world's finest amplifiers.
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[^3]
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## LT60I

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20,000 orp.v. Overload pr
tection. Slide swritch elector tection. Silde switch selector $0 / 25 / 2.6 / 10 / 50 / 250 /$
1000 V . w. C. $0 / 10 / 50 / 250 /$ 1000 V . A.C. $0 / 50 \mu \mathrm{~L} / 25 /$
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 Model fret class versatile atrument mannfactured 13.8.8.R. to the highent 10/50/250/500/1000v.
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OUR PRICE $25 \cdot 97$. P. \& P. 25 w

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MODEL PL436
20 k Q/Volt D.C. $8 \mathrm{k} \Omega /$ H/3/12/30/120/600 b.C. $3 / 30 / 120 / 600$ A.C. $50 / 600 \mu \mathrm{~A} / 60 / 600$ mA. $10 / 100 \mathrm{~K} / 1 \mathrm{Meg} / 10$ $\mathrm{Meg} \Omega=20$ to +46 db . \&6.87. P. \& P. 12p.

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30,000 O.P.V. with over-
losd protection mirror acale
$0 / 5 / 2 \cdot 5 / 10 / 25 / 100 / 250 / 500 /$ 1.000v. D.C. $0 / 2 \cdot 5 / 10 / 251$ $100 / 250 / 500 / 1,000 \mathrm{~V}$. n/50 $\mu \mathrm{A} / 5 / 50 / 500 \mathrm{~mA}$.
amp. D.C. $0 / 60 / \mathrm{K} / 6 \mathrm{Meg}$ ${ }_{\text {a }} 0 \mathrm{Mp}$ Meg $\Omega$ 88.87. Post pald

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$\mathrm{A} . \mathrm{C} .0-30 \mu \mathrm{~A} / 300 \mathrm{~mA} .0-3 \mathrm{~K} / 30$ A.C. $0-30 \mu \mathrm{~A} / 300 \mathrm{~mA} .0-3 \mathrm{~K} / 30$
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$0 / 50 \mathrm{uA} / 1 / 10 / 100 \mathrm{~mA} / 1 / 10 \mathrm{Amp}$ D.C.
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## U4312 MULTIMETER

electrical use. 667 o.p.v. 0/3/1.5/7.5/30/60/150/300/ $600 / 900$ VDC and 75 mV $0 / 3 / 1.5 / 7.5 / 30 / 80 / 150 / 800$ $600 / 900$ VAC.
$0 / 300 / 4.41 \cdot 6 / 6 / 15 / 60 / 150$ 0/1-6/6/15/60/150/万00MA. 1.5/6 AMP. AC. $1 / 200$ A/3K $/ 30 \mathrm{~K}$ Accuracy DC $1 \%$. AC
Accuracy DC $1 \%$. AC L-J\%
 with sturdy metal carrying case. lealls and instructions. $29 \cdot 50$. P. \& P. 25 p .
HIOKO MODEL 700X
100000 O.P.V. Overlaad 100,000 O.P.V. Overload
protection. Mirror scale. protection. Mirror scale. $120 / 300 / 600 / 1200 \mathrm{~V}$ DC $1200 / 6 / 12 / 30 / 60 / 150 / 300 / 6011$ $15 / 30 \mu \mathrm{~A} / 3 / 6 / 30 / 60 / 150 / 300 \mathrm{~mA}$
$6 / 12 \mathrm{AMP} . \mathrm{DC} .2 \mathrm{~K} / 200 \mathrm{~K} / 2$ $6 / 12 \mathrm{AMP}$. DC. $2 \mathrm{~K} / 200 \mathrm{~K} /$
Meg 20 Meg ohm -20 to
 $\underset{\text { MOLAnt }}{\text { MO }} \mathbf{7 0 8 0}$ mirror

$-20 .+62 \mathrm{db}$.
$816 . \mathbf{P} . \&$ P. 25


HTIOOB4 MULTI-METER Features A.C. current rangea.
100,000 o.p.v. Mirror Scale, Overload protection. $0 /-5 / 2 \cdot 5 / 10 / 50 / 250 / 500 / 1000$ V D.C.
$0 / 2 \cdot 5 / 10 / 50 / 250 / 1000$ S. A.C. $0 / 20 / 250 / \mathrm{LA} / 2 \cdot 6 / 26 / 250$
10 Amp. D.C. 10 Amp. D.C. $0 / 20 \mathrm{~K} / 200 \mathrm{~K} /$ $2 \mathrm{MEG} / 20 \mathrm{MEG} .-20+62 \mathrm{~dB}$
12.50. P \& P. 25p.

KAMODEN 72.200
MULTITESTER
High
200, 200,000 o.p.v. Overload pro-
tection tection. Mirror acale. Ranges:
$0 / 06 / \cdot 3 / 3 / 30 / 120 / 600$ $0 / \cdot 06 / \cdot 3 / 3 / 30 / 120 / 600 /$
$1 \cdot 00 \mathrm{D} / \mathrm{D}$ / $1213 / 12 / 60$ $0 / 3 / 12 / 60 / 300 / 11,200 \mathrm{~V}$
A.C. $0 / 6 \mu \mathrm{~A} / 1-2 \mathrm{~mA} /$
$600 \mathrm{~mA} / 12 \mathrm{~A} . \mathrm{D} . \mathrm{C}$ !
$0 / 12 \mathrm{~A} . \mathrm{A} . \mathrm{C}$.
-20 to +63 dB.
200 meg ohms $0 / 2 \mathrm{~K} / 200 \mathrm{~K} /!\mathrm{meg} /$
TMK LAB TESTER.
100,000 O.P.V. 61 it
8cale Buzzer Short Clr-
cuit Check. Sensitivity cuit Check. Sensitivity: Volt A.C. D.C. Voles: Volt A.C. D.C. Volts V. A.C. Volts: 3, 10, 25 .
50. 250, $500,1,000 \mathrm{~V}$.
10. 100 .
$10,100,500 \mathrm{~mA}, 2 \cdot 5,10 \mathrm{gm} \mathrm{\mu}$. Resistance:
$1 \mathrm{~K}, \quad 10 \mathrm{~K}, 100 \mathrm{~K}, 10 \mathrm{MEG}, 100 \mathrm{MEOS}$ $1 \mathrm{~K}, \quad 10 \mathrm{~K}, 100 \mathrm{~K}, 10 \mathrm{MEG}, ~ 100 \mathrm{MEGR}$.
Decibels: -10 to +49 db , Plastic Case with Carring Handle. Size: 7hin. $\times 6^{\frac{z}{2} \text { in. } \times}$ 31 in . 818.95 . P. \& $P$. 25p.

## Model S-I00TR MULTIMETER

## TRANSISTOR TESTER

overload o.p.r. Mirror scale/ overload protection.
$-6 / 3 / 12 / 30 / 120 / 600$
v -6/3/12/30/120/600 $0 / 6 / 30 / 120 / 600$. V AC
$600 \mu \mathrm{~A} / 12 / 300 \mathrm{~mA}$ AC. O/12 DC. $0 / 10 \mathrm{~K} / 1$ MEG/100MEG. -20 to $+50 \mathrm{db} .0 \cdot 01-2$ MFD. Transistor tester measuren Alphs, beta and Ico. Complete With batteries, Instruction ntil leadm. 218-50. P/P 25p.


KAMODEN HM. 350 TRANSISTOR TESTER High quality instrument to test Reverge Leak
current and DC current. Amplification factor of NPN, PNP, transiators, diodes, SCR's etc. $4^{\prime \prime} \times$ $4!^{\prime \prime}$ clear scale meter. Operates from internal batterles. Complete with instructions, leads and carrying handle. $\$ 12.50$
Post 30 p.
 LI $10-60$ - 500 , operation E17-50. Post 25p.
MODEL U4311 SUB-STANDARD MULTI. RANGE VOLT AMMETER Sensitivity 330 ohms/Volt
AC and DC. $\begin{array}{lll}\text { AC and } & \text { DC. Accuracy } \\ 5 \% & \text { D.C. } & 1 \% \\ \text { AC. Bcale }\end{array}$ length 165 mm . $0 / 300 / 760 \mu \mathrm{~A}$ / $1.5 / 3 / 7.5 / 15 / 30 / 75 /$
$150 / 300 / 760 \mathrm{~mA} / 1.5 / 3 /$ $15 \mathrm{EAMPDCD} / 3 / 7.5 / 15 /$ $30 / 75 / 150 / 300$

$750 \mathrm{~mA} / 1.6 / 3 / 7.5$ | AMP AC O |  |  |
| :--- | :--- | :--- |
| 300 | 750 m | 75 |
| 150 |  |  |

 $0 / 750 \mathrm{mv} / 1-5 / 3 / 7.5 / 15 / 30 / 75 / 150 / 300$ plete with teat leads, manual and teat certacates. \$49-00. Post 50 p .
TE-65 VALVE VOLTMETER $\begin{array}{lll}28 & \text { ranges. } & \text { D.C. volta } \\ 1 \cdot 5-1,500 \mathrm{v} . & \text { A.C. } & \text { volta }\end{array}$ $1 \cdot 5-1.500 \mathrm{v}$. A.C. volta
$1 \cdot 5-1.500 \mathrm{v}$. Resigtance up to 1,000 megohms. 200/ 240 v . A.C. operation. Complete with probe and instructions. el7.50. $P$ P.
Additional
probes Adistional probes aval \&8.50. R.F. 24.12t, H. KAMODEN HM. 720B F.E.T. V.O.M. Inpat Impedance 10 meg Ranges:
Ranges:
$0 /-25 / 1 / 25 / 10 / 50 / 1$ $250 / 1000$. D.C. 1000 V . A.C. 0/25 $/ 2$ / $25 / 25 / 250$ MA D.C.

$$
-20 \text { to }+62 \mathrm{~dB}
$$

 $0 / 5 \mathrm{~K} / 50 \mathrm{~K} / 500 \mathrm{k}$
500 meg ohms. E14.95. Post 30 p

## A.C. VOLTMETER

10 meg. input 10 rangen: $01 /-003 / \cdot 1 / 3 / 1 / 3 / 10 / 30 / 100 /$ $300 \mathrm{~V} . \mathrm{R} . \mathrm{M} . \mathrm{S} .4 \mathrm{cps} .-1-2 \mathrm{Mc} / \mathrm{s}$. Decibels -40 to +bods. with leads and inatructions. Operation 230 V. A.C. $817 \cdot 50$. Carr. 25p.
THE MODEL 117 F.E.T. ELECTROMIC VOLTMETER Battery operated, I meg input, 26 ranger. Large $41^{\prime \prime}$ mirror вcaie. Bize 51"×44"×21";
DC VOLTS 0.3-1200v AC VOLTS $3-300 \mathrm{~V}$ RMS. 8.0-800V P-P. DC CURRENT - 12-
12MA. Resistance up


Complete with. Decibelh $-20 \mathrm{tu}+51 \mathrm{~dB}$. P. \& P. 20p. V.O.M.

Input finpedatice 10 mink
 $120 / 800 \mathrm{~V} . \mathrm{DC} 0 /$.3 /
$12 / 60 / 120 / 600 \mathrm{~V} . \mathrm{A} . \mathrm{S}$ $0 / 120 \mu \mathrm{~A} / 120 \mathrm{~mA}$ D. $0 / 1 \mathrm{~K} / 100 \mathrm{~K}$
100 meg ohis 1100 meg ohund
515.97 . Post


KAMODEN HMG-500 INSULATION RESISTANCE TESTER
Range $0-1000$ Meg-
olung, 500 Voli. olms, 500 Vatti. Wide range clear meter $4 l^{\prime \prime} \times 4^{4 \prime}$. luxe carrying case batterles, instruc tions. $\mathbf{2 1 9}$.96. Post 301

BELCO AF-5A SOLID STATE SINE
SQUARE WAVE C.R. OSCILLATOR Sine $18 \times 200,000 \mathrm{II} \%$ : stumare $18 \times 50,000 \mathrm{~Hz}$
 tivit-1.95: 110 R . AMP. Bantwidh $500 \mathrm{~K} \mathrm{H}_{2}$ Sensitivity at 100 KHz , $\mathrm{H} 1 \mathrm{sis} / \mathrm{mmu} . ~ 3-25$ : Preset triggered sweel $: \cdot 3,000 \mathrm{usec}$; free running $20-200.000 \mathrm{~Hz}$ in rine rangen. Callbrator plps. $220 \times 360 \times 430 \mathrm{~mm}$. 115-230V. AC operatton. $238 \cdot 00$. C'arr. painl.

TO-3 PORTABLE OSCILLOSCOPE
3nn. tube, Y anup. Bensitucits $0.1 \mathrm{~V} \quad \mathrm{~F}-\mathrm{p} / \mathrm{CM}$. Bandwldth
1.5 cpg -1.5 MHz . Input imp

 Enput inip $\because$ mer $\Omega 20 \mathrm{pF}$ Time base. $\overline{5}$ radgen 10 cpp 300 kHz . Aytuchronization
Internalezterual. 1lluminate
 scale $140 \times 215 \times 330$ nim1. Weight lo 4 ll . handbook. \&47.50. Carr. 50 p .

## RUSSIAN CI-16 DOUBLE <br> BEAM OSCILLOSCOPE

$5 \mathrm{mc} / \mathrm{s}$ Pabs Band. Reparair
Rectang sitar 5 sin. $\times$ smplifiers.
C.R.T. Callbrated trig.
gered aweep from u/ser.
to 100 mllilinsec . per cmi. the Free running time base $50 \mathrm{c} / \mathrm{s}-1 \mathrm{Im} / \mathrm{s}$. Huitt.
in time base callbrator and amplitude in time base calibrator and amplitude accessories and inatruction mammal ع8\%. Carr. Pald.

## MODEL AT201

## DECADE ATTENUATOR

 Frequency ranke $u$ 200 KHz . Attenuator $0-111 \mathrm{db}, 0.1 \mathrm{db}$ step.Impedance 600 thma. Max. Input power


30 dbm . Slize $180 \times 90$
ARF- 300 AF/RF SIGNAL GENERATOR

TE－20D RF SIGNAL GENERATOR Accurate wide range yigual generator covering 120
$\mathrm{Kc} / \mathrm{g} 500 \mathrm{Mc} / \mathrm{s}$ on 6 band Kc／a $600 \mathrm{Mc/s}$ on 6 bands． Directly calibrated Vari－ able R．F．attenuator，audiu
output．Xtal socket fur callbration． $220 / 240 \mathrm{~V}$ ．A．C Brand new with instrue． tione el5 Carr．371 Bize $140 \times 215 \times 170 \mathrm{~mm}$


TE22 SINE SQUARE WAVE AUDIO GENERATORS 8tne：20cps to 200 8quare：20cps to 30 kc／s．Output impe． $200 / 250 \mathrm{~V}$ ．A．C．拉． eration．supplieil rand new alit truction marua and leads． 217.60
Carr． 87 \＄p

## ＂YAMABISHI＂VARIABLE VOLTAGE TRANSFORMERS Excellent quality at low cost．All dutels－

 Input


## PS． 200 REGULATED P．S．U．

PS．200 REGULATED P．S．U．
Bolld state．Variable output
5.20 volt D．C．up to 2 anop． 5．20 volt D．C．up to 2 arup． Independent meters to
monitor voltage and cur＊

 \＆19－96．P．\＆P．25p．


## SEW CLEAR PLASTIC PANEL METERS

USED EXTENSIVELY BY INDUSTRY，GOVT．DEPTS．，EDUCATIONAL
AUTHORITIES，ETC．
Over $\mathbf{2 0 0}$ ranges in stock－other ranges to order．Quantity discounts available． Send for fully illustrated brochure．

| Type SW． $100100 \times 80 \mathrm{~mm}$ ． |  |  |  |
| :---: | :---: | :---: | :---: |
| 50 | 24.15 |  |  |
| ${ }^{50-0-50} \mu \mathrm{~A}$ | ${ }^{23} \cdot 95$ |  |  |
| ${ }_{100-0-100 \mu \mathrm{~A}}^{100}$ |  |  |  |
| $600 \mu \mathrm{~A} \quad \cdots \mathrm{l}$ | 83．70 |  |  |
| 1 mA | 48．60 |  |  |
| 20V．D．C． | 23．80 |  |  |
| s0v．b．c． | 83.60 | 5 amp．I．c． |  |
| 300 V ．D．e． | E3．60 | 300 V ．A．C．． | 28.70 |
| 1 an！p．D．e． | 83．60 | ＇U Meter | 84．30 |
| Type $50.83082 .5 \mathrm{~mm} \times 110 \mathrm{~mm}$ Fronts |  |  |  |
|  |  | 10 ma | 83.10 |
|  |  | 80ma | \＄3．10 |
|  |  | 100 ma A | ${ }^{23} 310$ |
|  |  |  | ${ }_{\text {¢3．}}^{\text {¢ }} 10$ |
|  |  | $\frac{1}{5 \mathrm{mmp} . .}$ | ${ }_{\$ 3.10}$ |
|  |  |  |  |
| $50 \mu \mathrm{~A}$ | 23.40 | bV．D．C． | 23．10 |
| 50－0－5 | 23．40 | 10V．D．C． | ${ }^{ \pm 3.10}$ |
| $100 \mu \mathrm{~A}$ | 23．35 | 20 V ．D．C． | ${ }^{\text {c3．}} 10$ |
| 100－0－100 $\mu \mathrm{A}$ | 23.35 | ${ }^{50 \mathrm{~V} .1 . \mathrm{C}}$ ． | E8．10 |
| $200 \mu \mathrm{~A}$ | 83.30 | $301 \mathrm{~V} . \mathrm{D.C}$ | 23.10 88.30 |
| ${ }^{500 \mu} \mathrm{~A}$ | 退3315 | ${ }_{300}^{15 \mathrm{~V} . \text { A．C．C．}}$ | ＋38．30 |
| ${ }_{\text {bind }}$ | 23．10 | v0 Meter | 28.50 |
| Type $50.64063 .5 \mathrm{~mm} \times 85 \mathrm{~mm}$ Fronts |  |  |  |
| 50ua | 43.05 | 50 ma | 22．90 |
| $50-0-50 \mu \mathrm{~A}$ | 23－05 | $1 \mathrm{kmp} .$. | 82.90 |
| $100 \mu \mathrm{~A}$ | 23.00 | 5 emp．． |  |
| 100－0－100 A A | E3．00 | 10 amp | E2．90 |
| $200 \mu \mathrm{~A}$ | 23.00 | ${ }^{8} \mathrm{~V}$ ．D．C． | 22．90 |
| ${ }^{600 \mu} \mathrm{~A}$ | 82．95 |  |  |
| ${ }_{5 \mathrm{ma}}^{1 \mathrm{ma}}$ | 22．90 | ${ }^{30} 0^{\text {a }}$ V．D．D．C． | 42.90 82.90 |
| 10 mA | 22－90 | 15 V A．C． | Es 00 |
| s0ma | 22.80 | ${ }^{309 \mathrm{~V} . \text { A．C．}}$ | ${ }^{28} 00$ |
| 100 mA | 82．90 | vU Meter | 23.15 |
| Type $50.46046 \mathrm{~mm} \times 59.5 \mathrm{~mm}$ Fronts |  |  |  |
| sous | 22．80 | 1 mmp. |  |
| 60－0－50 4 A | 22－80 | 5 amp． | 22． 60 |
| $100 \mu \mathrm{~A}$ | 22．75 | 10 amp． |  |
| 100－0－100 A A | 22.75 | SV．D．C． | \％280 |
| $200 \mu \mathrm{~A}$ | 28.70 | $10 \mathrm{~V} . \mathrm{D.C}$. | 22.60 82.80 |
| $500 \mu \mathrm{~A}$ |  | 20\％．D．C． |  |
| ${ }_{50 \mathrm{ma}}^{1 \mathrm{ma}}$ | 82．60 | 80\％．D．C． | 22．60 |
| 5 sma |  | $306 \mathrm{~V} . \mathrm{D.C}$. | 28．60 |
| 50 mA | E2．60 | $15 \mathrm{~V} . \mathrm{A} . \mathrm{C}$ ． | 22．70 |
| 100 mA | 82－60 | 300 V ．A．C． | 22．70 |
| 500 mA | 22－60 | VU Meter | 22.90 |


＂SEW＂EDGWISE METERS Troe PE．70． 3 17／32in．x 1 15／32in．$x$


## －MOVING IRON－

ALL OTHERS MOVING COIL
Please add pcstage


|  | 50 ma | 88. |
| :---: | :---: | :---: |
|  | 100 mA | e8 |
|  | 500 mA | 88 |
|  | 1 amp ． | ¢ 8 |
|  | 5 amp ． | £3 |
| ， | 15 amp ． | £3 |
|  | 30 smp | 48 |
|  | 20V．D．C． | 13 |
|  | 50 V ．D．C． | 8 |
|  | 150 V ．I．C． | 88 |
|  | 300 V ．D．C． | 8 |
|  | 15 V ．A．C． | 8 |
|  | 300 V ．A．C． | \＆ |
|  | S Meter lma | 28 |
|  | VU Meter | \＆ |
| 0 | 1 amp．A．C．＊ | 2 |
| 0 | 5 amp A．C．＊ |  |
| 0 | 10 mmp ．A．C．＊ | 48 |
| 0 | 20 amp ．A．C．＊ |  |
|  | 30 mmp ．A．C． |  |


| Type MR．52P． 2 tin．square fronts |  |  |  |
| :---: | :---: | :---: | :---: |
| $\mathrm{b}_{0} \mathrm{H}$ A | 29－50 | 10V．D．C． | E2．50 |
| 50－0－50 $\mu \mathrm{A}$ | 28．05 | 20 V ．D．C． | 22.50 |
| $100 \mu \mathrm{~A}$ | \＆ 3.00 | 50 V ．D．C． | 28．50 |
| 100－0－100 $\mu \mathrm{A}$ | 29．85 | 300 V ．D．C． | 48.50 |
| $500 \mu \mathrm{~A}$ | 82.65 | $15 \mathrm{~V} . \mathrm{A}$ C． | 28.60 |
| 1 mA | 22．50 | 300 V ．A．C． | 82.60 |
| Sma | 22.50 | 8 Meter 1 mA | 82.60 |
| 10 mA | E2．50 | vU Meter | 28.60 |
| 50 mA | 58－50 | 1 amp ．A．C． | 22.50 |
| 100 mA | 29.50 | 5 amp ．A．C． | 48.50 |
| 500 mA | 82．50 | 10 mmp ．A．C． | ＊ 22.50 |
| 1 amp | 22．50 | 20 mmp A．C． | － 82.50 |
| b amp | 28．50 | 30 mmp ．A．C． | ＊2．50 |

## Type MR．65P．3tin．x 3tin．fronts



20


10
10
5
1
1
1
1
2

1
5
10
15

| amp． | 82.60 |
| :---: | :---: |
| 10 amp ． | $42 \cdot 60$ |
| 15 mmp ． | 82.60 |
| 20 amp ． | 82.60 |
| 30 mmp ． | 42.80 |
| 50 amp ． | 42.90 |
| D． | 22－60 |




| Type MR．45P． 2 in ．square fronts |  |  |  |
| :---: | :---: | :---: | :---: |
| $50 \mu \mathrm{~A}$ | 22．70 | 6 amp |  |
| 50－0－50 A | 22.65 | 10 V. D．C． |  |
| $100 \mu \mathrm{~A}$ | 82.60 | 20 V ．D．C． | E2 |
| 100－0－100 $\mu \mathrm{A}$ | 22.50 | 50 V ．D．C． | 2 |
| $200 \mu \mathrm{~A}$ | 42.50 | 300 V ．D．C． |  |
| $500 \mu \mathrm{~A}$ | 22.45 | 15V．A．C． |  |
| $500-0-500 \mu \mathrm{~A}$ | 22.40 | 300 V ．A．C． |  |
| 1 mA | 22.40 | $\underline{\mathrm{S}}$ Meter $\operatorname{lm} \mathrm{A}$ |  |
| 5 mA | 42.40 | $V \mathrm{U}$ Meter | 4 |
| 10 mA | 22.40 | 1 amp ．A．C． | e2 |
| 50 mA | 22.40 | 6 amp ．A．C． | ＊2 |
| 100 mA | 82.40 | 10 mmp ．A．C | 82 |
| 500 mA | 22.40 | 20 amp A．C． | 22 |
| 1 amp ． | 22.40 | 30 amp ．A．C． | 82 |


| ＂SEW＂BAKELITE <br> PANEL METERS |  |
| :---: | :---: |
| Type MR．65． $3 \frac{1}{1} \mathrm{in}$ ．square fronts |  |
|  | 1 amp．．．．．．． 28.60 |
|  | $5 \mathrm{amp} . . . . . . .88 .80$ |
|  | $15 \mathrm{amp} . . .$. |
|  | 30 amp，．．．． 88.60 |
|  |  |
|  | $\begin{array}{lll} \text { SV. D.C. } & 22 \cdot 60 \\ \text { 10v. D.C. } & . & 22.60 \end{array}$ |
|  | 20V．D．C．． 88.80 |
|  | 50 V. D．C．．．82．60 |
|  | 150 V ．D．C． 82.60 |
|  | 300 V ．D．C．$\quad 22.60$ |
| $25 \mu \mathrm{~A}$ ．．．．． 44.60 | 50 mV ，D．C．$\quad 22.90$ |
| 504．A ．．．．． 88.55 | 100 mV ．D．C． 82.90 |
| $50-0-50 \mu \mathrm{~A}$＊3．05 | 30 V ．A．C．＊．． 22.65 |
| $100 \mu \mathrm{~A}$ ．．．． 28.00 | 50 V．A．C．${ }^{\text {¢ }}$ ： 28.65 |
| 100－0－100 4 A － 88.00 | 150V．A．C． 82.65 |
|  |  |
| 800－0－500 4 A | $500 \mathrm{~mA} \mathrm{A.C} \quad$. |
|  | 1 mmp A．C．${ }^{2} 2.60$ |
|  | ${ }^{5} \mathrm{mmp}$ ．A．C．${ }^{\text {c }}$ 2． 60 |
| 5 mA ．．．．．． 2260 |  |
| 10 mA ．．． 28.80 | 20 amp ．A．C．＊${ }^{2} \mathbf{2} \cdot 60$ |
| boma ．．．． 28.60 | 30 amp．A．C．${ }^{\text {2 }}$ 2． 60 |
| 100 mA ．．． 28.60 | 50 mmp A．C．＊ 28.60 |
| 500 mA ．．．． $22 \cdot 60$ | VU Meter ．．88．65 |


| Type S－80 80 mm ．square fronts |  |  |
| :---: | :---: | :---: |
| 50\％A | 28．50 |  |
| $50-0-60 \mu \mathrm{~A}$ | 28.40 |  |
| $100 \mu \mathrm{~A}$ | 88.40 |  |
| $100-0-100 \mu \mathrm{~A}$ | 88.80 |  |
| $500 \mu \mathrm{~A}$ | E8．05 |  |
| 1 mA | 88.00 |  |
| 20v．D．C． | E3．00 |  |
| 50V．D．C．． | 28.00 | 5 arap．D．C． 88.00 |
| 300 V ．D．C． | 88.00 | $300 \mathrm{~V} . \mathrm{A.C}$. ． 28.00 |
| 1 smp. D．C． | 88.00 | VU Meter ．．\＄3．70 |

## POWER RHEOSTATS

 High quality ceramic con－ truction．Windinge em－ bedded in vitreous enamel．Heavy duty brush wiper． Continuous rating．Wide range avallable ex－Rtock． Single hole fixing．lin．dia． shafts．Bulk quantities avail． sble．
25 WATT．10／25／50／100／250／500／1000／1500／ 2500 or 5000 ohms． 96 p. P．\＆P． 10 p ． 50 WATT．10／25／50／100／250／500／1000／2500 or 5000 ohms．$\$ 1 \cdot 85$ ．P．\＆P． 10 p ．
100 WATT． $1 / 5 / 10 / 25 / 50 / 100 / 250 / 500 / 1000$ or 2500 ohms． 21.96 ．P．\＆P． 15 p ．

$240^{\circ}$ Wide Angle ImA Meters MW 1.6 60mm $\begin{aligned} & \text { 日quare } 23.97\end{aligned}$ P．\＆P．extra

A.C. Bran UR-IA RECEIVER


4 Bands covering $550 \mathrm{kc} / \mathrm{s}-30 \mathrm{mc} / \mathrm{s}$ FET, \& Meter. Variable BFO for S8b. Built-in Speaker, Band; spread, Bensitivity Control. $220 /$
240 v . A.C. or 12 V . D.C. $124 \mathrm{in} . \times$
 $43 \mathrm{in} . \times 7 \mathrm{in}$.
instructions.

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$500 \mathrm{mF} .12 \mathrm{~V}, 15 \mathrm{p} ; 25 \mathrm{~V} .20 \mathrm{p} ; 50 \mathrm{~V} .80 \mathrm{p}$.
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70 p. $1000 \mathrm{mF} .12 \mathrm{~V} .17 \mathrm{p} ; 26 \mathrm{~V} .85 \mathrm{p} ; 50 \mathrm{~V} .47 \mathrm{p} ; 100 \mathrm{~V} .70 \mathrm{p}$ ${ }_{2500 \mathrm{mF}}^{2000 \mathrm{mF} .} 60 \mathrm{~V} .26 \mathrm{p}$; 28000 mP . 25 V . $47 \mathrm{p} ; 50 \mathrm{~V}, 65 \mathrm{p}$. 5000 mF . 6V. $25 \mathrm{p} ; 12 \mathrm{~F} .42 \mathrm{p} ; 25 \mathrm{~V} .75 \mathrm{p} ; 85 \mathrm{~V} .85 \mathrm{p} ; 50 \mathrm{~V}$. 05 p .

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DURING the first 18 years of human life a great many natural developments take place which perhaps we take for granted. One of these is the shrugging off of juvenile behaviour characteristics and the adoption of a mature outlook to face the rigours of a competitive world.
When applied to a technical art such as broadcast entertainment, not only do programme standards alter, but also the medium by which they are conveyed. The obvious examples that spring to mind are the progressions through 405 -line, 625 -line, and colour television for one; acoustic gramophones, mono discs, 2-channel stereo and quadraphonic sound for another. In spite of a mixture of pessimism, cautiousness, apathy, or just plain rejection, developments in the methods and standards of technical achievement still win through.
But there is still a very interesting factor; that is a great reluctance, especially in the U.K., to shake off some of those formative habits. We still have 405 -lines, mono discs (often forced to undergo processing to stereo status), and of more topical concern, medium wave broadcasting. Why? Or equally, why not? The very organisation that has been trying to convert us to v.h.f./f.m. for the past 18 years is still promoting broadcasting on m.w./l.w. Even more interesting-the IBA are now doing likewise. Why?
Have you bought or built an f.m.-only tuner in the light of the advice about improved reception of BBC programmes, only to find that you cannot receive the programmes that go out on a.m. only? Do you also find that reception in cars via f.m. is very difficult because of interference? Are you finding station separation on the medium wave more difficult? It is truly amazing that in spite of "progress" our broadcasting system is still bugged on occasions by major reception problems. (See letters on page 261.)
It is interesting to note that the BBC are experimenting with concurrent television and stereo radio programme material. Let's have more! Experiments are also going ahead on broadcasting quadraphonic sound, but we have not heard them yet.

In the meantime we will go on promoting v.h.f./f.m., especially for constructors. Mr. Ainslie's recent article was typically a popular project. Mr. Rayer describes a special receiver in this issue, that enables the experimenter to extend reception outside the usual $88-108 \mathrm{MHz}$ limits of a conventional dial. Future issues will provide state-of-the-art circuits of f.m. stereo. But we will not forget a.m. On the contrary we shall do all we can to provide constructors with limited experience some simple circuits to build, and at low cost too; more details about this next month.

## M. A. COLWELL-Editor

The August issue on sale 6th July, will include our new all-transistor oscilloscope specially designed for students, experimenters and constructors who want reliable performance coupled with simple construction. We shall also have a communications receiver and details of a new Datacard scheme commencing in the autumn. Our autumn issues are being planned as the best yet for new projects and ideas on home entertainment. We would hate the idea of you missing them, so order your regular copies of Practical Wireless now.

Further details on page 239.

## Mobile rally diary

June 17-Amateur Radio Mobile Society Rally, R.A.F. Cosforú, Shropshire. Details from W. S. Barwick, 34 Malvern Rd., London. N. 8.

June 24-West of England (Bristol City and County RSGB Group) Mobile Rally at Longleat. Further details of this popular rally can be obtained by dropping a line to A. H. Williams, G8CKJ, 58 Britannia Road, Kingswood, Bristol.
July 1-South Shields and District Amateur Radio Club's rally will be at Redwell School, Prince Edward Road, South Shields.
July 8-Cornish Rally at Treviglass County Secondary School, Newquay, Cornwall.
July 8-The Worcester and District Amateur Radio Club are organising the Upton-on-Severn Mobile Rally and further details can be obtained from B. A. Jones, G8ASO, 12 Woodside Road, Larkhill, Worcester, WR5 2EG.
July 22-Anglian Mobile Rally will be held at the Suffolk Showground Bucklesham Road, Ipswich. Readers can obtain further information by contacting $C$. J. Wantling, G3TNE, 15 Melplash Road, Ipswich, Suffolk, IP3 8QJ. Telephone Ipswich 75241.
July 22-Southdown Rally at Polegate, Wilmington.
August 5-Radio Society of Great Britain Mobile Rally at Woburn Abbey, Beds.
August 12-The Derby and District Amateur Radio Society are holding their 1973 rally at the usual venue i.e. Rykneld School, Bedford Street, Derby. There's ample inside accommodation if it rains and plenty of parking space for cars. In addition, there will be plenty of attractions for the XYL's and harmonics. If you want to know more, send a s.a.e. to Fred Ward, G2CVV, 5 Uplands Avenue, Littleover, Derby, DE3 7GE.
August 12-Torbay Annual Mobile Rally will be held at Newton Abbot Rugby Club ground. There will be talk-in stations, the usual stands and various competitions. The contact for further gen is L. H. Webber, G3GDW, 43 Lime Tree Walk, Newton Abbot, Devon.

## An Aussie 'Eirst'

RON Wilkinson, VK3AKC is the first man in Australia to send a radio signal to America via the moon. He used the frequency of 1296 MHz and the signal was received by the Crawford Hills UHF Group U.S.A.

## MOS i.c.'s and their Applications

THE latest book published by Mullard Ltd. is entitled "MOS Integrated Circuits and their Applications". It provides the reader with an introduction to these devices and should be ideal reading for designers unfamiliar with them and for electronic engineering students.

The construction, manufacture and characteristics of MOS i.c.'s are covered and the ways in which they can be used are fully explained.

A short description of the operation of the devices is followed by the explanation of the manufacturing processes of the various types of MOS i.c.'s. The static characteristics of the MOS transistor are examined and a description given of the static inverter which forms the basis of the gating elements used in MOS logic devices.

Switching characteristics are discussed and the main MOS static and dynamic logic circuits are examined.

Interface circuits are given and input/output protection devices described.

The final chapter deals with typical logic applications in which MOS i.c.'s can be used advantageously.

The book, cloth bound with 176 pages costs $£ 2$ - and is obtainable from leading booksellers.

## Interface '73

TEXAS Instruments will again be holding their annual Seminar at the 'Talk of the Town' in London. It will be entitled "Interface 73" and will run for 3 days-June 12, 13 and 14. The same programme will be repeated each day.

In addition to presentations on Optoelectronics, Memories and Power Devices etc. Texas Instruments are expecting to make important policy statements on the Digital I.C. market.

## Audio Fair '73

DATES are announced for the 1973 International Audio Festival and Fair and, for the first time since the exhibition moved to Olympia in 1969, it is to be open on a Sunday. Dates are October 23 to 28.


## Circular aerials for VHF

EMI has been awarded contracts to provide the Independent Broadcasting Authority with transmitting aerial systems for Britain's first commercial radio stations.
Altogether six systems are being supplied and they will cover five major cities throughout the country including London, Birmingham, Glasgow and Manchester. Transmissions from the new stations are expected to commence towards the end of 1973.

Two v.h.f. aerials have been specified for London and Birming. ham, which will be the first of their type in this country, utilising circular polarisation techniques. These techniques are used extensively in the United States to give improved reception for transistor and car radios.

The aerial system for London, which is 60 ft . high, will be installed on the top of the 500 ft . television tower at the IBA Croydon transmitting station. This aerial will transmit two programmes to the London area. The second v.h.f. system will be mounted on the 1000 ft . mast at the Lichfield transmitting station, near Birmingham



## CIRCUIT

Fig. 1 is the circuit of the receiver, the rod aerial or feeder being taken to the primary of L1, which is tuned by VCl. L2 is the oscillator coil, capacitor VC2 having a ball drive and dial for tuning. L1 and L 2 are miniature plug-in coils, for changing bands.

Trl is the dual-gate FET mixer, output from the drain going to IFT1. Transistors $\operatorname{Tr} 3, \operatorname{Tr} 4$ and $\operatorname{Tr} 5$ are the three 10.7 MHz IF amplifiers, with doubletuned IF transformers. C8 and C11 are necessary to neutralise these stages. IFT4 is a ratio discriminator transformer with the matched diodes D1 and D2 the audio output being taken via R19 from the tertiary winding.

VR1 is the audio volume control feeding a 3-stage amplifier. A $21_{2}$ in. speaker is incorporated for portable use, but the output jack allows a larger speaker or headphones to be plugged in.

The mixer and oscillator stages, with associated components, are built on a metal chassis but the IF and AF amplifiers are built as separate units on insulated boards.

## CHASSIS AND PANEL

The chassis is $8 \times 4 \mathrm{in}$. and is a universal chassis flanged member. For the case described, the panel is flat aluminium, $8 \times 5 \mathrm{in}$. Fig. 2 shows the underside of the chassis.

VCl is mounted centrally and ${ }_{1}{ }_{4} \mathrm{in}$. clear of the chassis. The holder for L1 is placed so that the leads from pins 1 and 6 are as short as possible, as shown.

VC2 is mounted on a metal bracket and is set back to accommodate the ball drive. The holder for L2 must also be placed to allow very short leads to VC2. The ball drive requires a ${ }_{4} \mathrm{in}$. or lin. hole in the panel and its lug is fixed with a 6BA bolt. VC2 and the drive need to be lined up to give a smooth free movement. The listed drive has a flange and

For general VHF use, Band 3 coils will be mos! suitable, the reception coverage then being approximately $72-130 \mathrm{MHz}$, with the oscillator HF of the signal frequency.
components list

| Resistors |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| R1 100ks | R13 | 680s? | R25 | 47kS |
| R2 2.2k | R14 | $1 \mathrm{k} \Omega$ | R26 | $3 \cdot 3 \mathrm{ks}$ ! |
| R3 100ks2 | R15 | 33 kS | R27 | 47ks2 |
| R4 $1 \mathrm{k} \Omega$ | R16 | $5 \cdot 6 \mathrm{k} \Omega 2$ | R28 | 12kS |
| R5 10ks | R17 | $1 \mathrm{k} \Omega$ | R29 | $2 \cdot 7 \mathrm{k} \Omega$ |
| R6 10ks | R18 | 680,2 | R30 | 680S2 |
| R7 33 kS 2 | R19 | 1008 | R31 | $2.2 \mathrm{k} \Omega 5^{\circ} \%$ |
| R8 $5 \cdot 6 \mathrm{ks} 2$ | R20 | 680S2 | R32 | $39 \Omega 5 \%$ |
| R9 680S2 | R21 | 680S2 | R33 | $5 \cdot 6 \mathrm{k} \Omega$ |
| R10 $1 \mathrm{k} \Omega$ | R22 | $4 \cdot 7 \mathrm{kS} 2$ | R34 | $470 \mathrm{k} \Omega$ |
| R11 33ks | R23 | 4.7kS2 | VR1 | $22 \mathrm{k} \Omega \log$. |
| R12 5.6ks | R24 | $5 \cdot 6 \mathrm{k} \Omega 2$ | pot. | with switch S1 |




Fig. 2: Layout underneath the chassis with the tuned oscillator stage at the right. The mixer stage board (Fig. 3) Is in the centre with its associated tuned circuit. The speaker is interconnected with the headphone jack socket mounted on the rear apron of the chassis.

## MIXER BOARD

This board is $17_{8} \times 3_{4} \mathrm{in}$. of $0 \cdot 15 \mathrm{in}$. matrix plain Veroboard, wired as in Fig. 3. A 6BA bolt, with extra nuts, allows the board to be locked to the metal chassis, with enough clearance to avoid any short circuits. This bolt forms the earth return.

The leads of C1 and C3 can be left projecting, to solder to VC1 and VC2. Also fit a pin or projecting lead for the negative connection, and a wire which can pass down through the board and chassis, to reach pin 3 of IFT1 on the IF amplifier.

Transistor Trl should be supplied with a shorting clip on its leads, and this must not be removed until the transistor has been soldered in place. First wire the board as in Fig. 3, omitting Tr1. Solder on the wires which will go to gate 1 , gate 2 , source and drain, and cut them to length and tin them. Trl is then inserted with its wires as shown. The leads can


Fig. 3: Both sides of the mixer stage board.
be spread with a sharpened matchstick, but should not be touched with a metal or plastic tool, or the fingers. Solder source and drain leads. Then while holding iron and solder in contact with the lead from Cl/R1, solder gl. In the same way, hold the iron and solder against the lead from R3, and solder g2. When the resistors are soldered to provide an external circuit from g1 and g2 to the source, the transistor is protected and the wire clip is removed. If gl or g2 have to be unsoldered from R1 or R3, or these resistors have to be unsoldered, the transistor leads should first be bound with thin wire.

The board is fixed as in Fig. 2, and external connections are made. The co-axial aerial socket and output socket are fixed to the rear flange of the chassis so that the receiver can be tested out of its case.

## IF AMPLIFIER

This is assembled on a $51_{2} \times 1{ }_{4}$ in. board as in Fig. 4. First mark and drill holes for the IFT's and for the two ${ }^{1}{ }_{2}$ in. 6BA bolts. These, with additional nuts, form the chassis return and hold the board the metal chassis.

Leads should be short and direct, and insulated sleeving is put on where necessary. If ${ }^{1}{ }_{4} \mathrm{in}$. lengths of small diameter sleeving are put on the transistor leads this will prevent shorts and hold the transistors at a suitable height, while also identifying the wires.

Note that all the IFT screening cans are earthed. As the IFT's are prealigned, unnecessary disturbance of the cores is better avoided. When wiring D1 and D2, note that these may have a black ring to show positive. Polarity must be correct as a DC voltage is produced across C21.

Run a lead from pin 3 of IFT1 to a pin or wire, so that the wire from Tr 1 can be soldered on later. A black lead, to go to battery negative, is soldered to R18, Fig. 4. A lead from R19 is provided, and will pass through the chassis to VR1.
If this section is tested with the board separate from the receiver, do not run a signal generator or other lead from pin 3 of IFT1 near the other IFT's or circuits, or instability may be caused which would not be present when the amplifier is fixed to the chassis.


View of the underside of the FM receiver which should be compared with Fig. 2.


Fig. 4: The two sides of the IF amplifier board, shown actual size, illustrating component layout and wiring.

## IF ALIGNMENT

This can be carried out with the aid of a highresistance voltmeter, or the 2.5 V range of the usual type of multi-range meter, preferably with a resistance of $10 \mathrm{k} \Omega / \mathrm{volt}$ or better. It is assumed that if alignment equipment for FM raceivers is available the means of employing this will be known, so the method of alignment described here is an alternate method.

Three test points are shown on the circuit, Fig. 1, and the multi-range meter is clipped to these for aligning the IF amplifier by this method. Initially, meter negative is clipped to $A$ and meter positive to B. Maximum signal strength at the diodes will then produce maximum voltage from $A$ to $B$, as shown by the meter. All the IFT cores are adjusted for maximum meter reading and each core should have quite a sharp tuning peak, seen by observing the meter while slowly turning the core.

For the second stage of adjustment, the meter is clipped from point C to chassis, the polarity being unimportant. The bottom core of IFT4 is then adjusted for minimum reading on the meter. In addition to the usual upwards movement of the meter pointer, some backwards movement can normally be seen. By rotating the lower core one way or the other, it should be possible to make the meter pointer move backwards or forwards from its minimum position. Set the core so that the meter shows this "zero". This completes the IF strip adjustment.

When making these adjustments, a properly fitting plastic trimming tool should be used. It is helpful to have a signal generator of the usual type employed with AM receivers and to tune this to $10 \cdot 7 \mathrm{MHz}$. A lead from the generator is placed near the receiver mixer, to secure sufficient input to allow an indication on the voltmeter as described. When the generator output is amplitude modulated, this will be heard in the speaker, but the adjustments described are made with the generator giving a CW output.

Unnecessary adjustment to the cores before obtaining results is best avoided, as they are approximately adjusted when the IFT's are manufactured. Should IFT's be used which are so far out of adjustment that no signal can pass through the IF strip, couple the signal generator to the base of the last IF amplifier and adjust the last IFT. Then transfer the generator input to the previous amplifier base and adjust the third IFT. After taking the input to the first IF amplifier base and adjusting the second IFT transfer the input to the mixer in the way described, and carefully check all the IFT's.


Fig. 5: Details of the audio amplifier board, full size.

Where no signal generator is available, the only method is to tune the receiver until some signal is obtained, and to work on this by ear, adjusting all cores for best volume and the bottom core for best results with an FM transmission. This can prove reasonably satisfactory, unless the IF circuits are so far off tune that the strongest transmission available cannot be heard.

## INSTABILITY

With three IF stages and double-tuned transformers it is quite possible for IF instability to arise. This depends on the individual transistors, position and length of leads and similar factors. It is heard as whistling or hissing accompanying signals, or as oscillation liable to start when the IFT's are aligned, but which ceases when some cores are slightly staggered.
Where a small adjustment of some cores prevent this, and results are good, no modification is necessary. Otherwise, the trouble can be cured by placing a resistor between pin 5 of one or more LFT's and the corresponding base lead. This is most easily done by disconnecting the base wire from the pin, so that the resistor can be soldered directly to the pin, under the board. A similar result is achieved by placing a resistor in the lead from collector to pin 3 of the following IFT.
The values of the resistors are not very critical, except that unnecessarily large values will reduce gain. A number of small resistors from about $22 \Omega$ to $220 \Omega$ or so can readily be tried. In the original IF strip a value of $100 \Omega$ was used in the first stage, with $39 \Omega$ resistors for each of the following stages.
Though the amplifier can be tested and roughly aligned when clear of the receiver, final adjustments should be made with it fixed in its normal position. Four holes are required in the metal chassis, so that the lower cores can be reached. Final slight readjustment of IFT4 can be made by ear, when a transmission is tuned in. A simple $10 \cdot 7 \mathrm{MHz}$ alignment generator is described later.

## COILS

Band 3 was covered with Denco miniature plug-in coils Range 7. A "Yellow" coil is used for L1 and "Red" coil for L2. This is the most generally useful band. Band 2 is covered by using a similar pair of coils, Range 6. The Range 5 coils, which have cores, cover Band 1. All these coils are the valve type. For Band 4, coils were wound using Denco plug-in formers. Place the former in a valve holder before soldering to the pins. L1 is $2^{1}{ }_{2}$ turns spaced ${ }^{1}$ gin. and tapped at 1 turn from pin 1, for pin 9. L2 is $1^{1}{ }_{2}$ turns, spaced ${ }^{1}{ }_{8}$ in. Both are 22 swg or similar wire.

## AF AMPLIFIER

This is assembled on $0 \cdot 15 \mathrm{in}$. matrix Veroboard, in a similar way to the IF amplifier. The board is $2^{1}{ }_{2} \mathrm{in}$. $x 2$ in. and both sides are shown in Fig. 5.
Transformer T2 has lugs which pass through slots, which can be made by drilling small holes closely together, and the lugs are twisted to hold the transformer to the board. Two ${ }^{1}{ }_{2} \mathrm{in}$. 6BA bolts secure tags which form the chassis (positive) return.

The wiring of this unit should prove to be quite straightforward. Observe the polarity of electrolytic
capacitors and put sleeving on any leads where this is necessary to avoid short-circuits. Two thin flexible leads are soldered to the secondary tags of T2, for the speaker connections. A Veropin or projecting wire is provided for the connection to VR1 and another for fixing a black flexible lead for battery negative.

A great advantage of this circuit is that it provides quite high gain while the DC operating conditions of each stage are isolated from those of other stages. The specified transistors are readily available, but if similar types are to hand it is quite likely that they could be used. In some cases, the value of one or more resistors might have to be changed to suit the transistors. This is especially so in the output stage, where unsuitable values may cause high battery drain on the one hand, or very low drain with distortion, on the other.

It is quite a good plan to test the AF amplifier by itself, when it is completed. This is done by taking an AF input to the volume control VR1, which may be left in circuit to avoid overloading. The speaker and battery are connected when the amplifier should draw about 10 mA , rising to $25-50 \mathrm{~mA}$ or so with ample volume.

The parts are assembled with glue and panel pins. Four ${ }_{4} \mathrm{in}^{2}$. strips are glued flush with the front edges, inside. The receiver can then be inserted as a complete unit from the back.

Bolts through the holes which are punched in the chassis flanges hold the receiver in the case. The back is thin plywood or paxolin, held by small woodscrews.

An extending telescopic rod aerial can be fixed to the left-hand side and connected by a short flexible lead to the co-axial socket. The latter can also be used for an external aerial.

## AERIALS

In those areas where VHF transmissions are normally well received a telescopic aerial or short wire should be quite adequate. The wire can be 2 ft . to 6 ft . or so long. The position of the receiver and aerial can influence volume, even with a vertical aerial.

It is also easy to try wires which are very long in terms of wavelength and part of the aerial system can then be out of doors. But with rather long wires the chances of $10 \cdot 7 \mathrm{MHz}$ breakthrough are increased.


This photograph shows the position of the IF and AF boards on top of the chassis.

The negative feedback is not essential and R34 may be omitted or changed to between 470 and $680 \mathrm{k} \Omega$. If used, connect it from driver base to one output transistor collector at T2 primary. If volume falls slightly, this is correct. Should oscillation begin, transfer the connection to the other collector.

As the first AF stage is followed by two stages of amplification, a noisy transistor here will cause a hiss in the speaker. This was so in the amplifier shown and was traced to the use of a noisy, cheap surplus 0C71 in the first stage. When this transistor was replaced, the noisy background ceased.

## CABINET

The cabinet is made from ${ }^{1} 4 \mathrm{in}$. plywood. Each side is $5^{1}{ }_{4} \times 4^{1}$ in. and the top and bottom are $85_{8} \times 4^{1}{ }_{2} \mathrm{in}$. An aperture to match the small internal speaker is cut in the right-hand side and covered with speaker gauze or similar material.

Best possible results are obtained by making a VHF type aerial of suitable dimensions for a chosen band. The aerial can then be high and in the clear outside, a co-axial down-lead going to the receiver. This can give a very great increase in signal strength.

## ALIGNMENT GENERATOR

If necessary, a simple $10 \cdot 7 \mathrm{MHz}$ alignment generator can be built using the circuit shown in Fig. 6. $\operatorname{Tr} 1$ is an audio oscillator and output from this may be taken, via the isolating capacitor Cl , from the "AF" pin. This output may be useful for checking the AF circuits as well as providing amplitude modulation (and a certain amount of FM also) of the RF oscillator Tr2. The RF oscillator is tuned by means of the core of L1 and output can be had via C5 at the "RF" pin. Enough input to the IF strip may be obtained by just placing the generator near the circuits of IFTI.


Fig. 6 : Circuit diagram of the simple alignment oscillator providing RF or modulated RF at $10 \cdot 7 \mathrm{MHz}$.

[^4]Constructing this generator is straightforward. A 5in. x 2in. flanged "universal chassis" member has its flanges cut 2 in . from one end, so that a $2 \times 2 \mathrm{in}$. section can be bent at right angles. Two toggle switches (Sl and S2) are mounted on this part, near the top and edges to clear the transformer Tl.

The generator is assembled on a small piece of Veroboard, as in Fig. 7. Drill the board for the pins of L1 and T1 and for two ${ }^{1}{ }_{2}$ in. 6BA bolts, which with extra nuts will provide the chassis return and allow the wired board to be mounted in the small chassis described. Two Veropins are inserted, at the points "AF" and "RF" so that leads can be clipped or soldered on, if wanted. If preferred, two small sockets could be mounted between the switches and be wired to these pins.

The circuit was found to operate with any battery from 3 V to 9 V . When S 2 is open, an audio tone accompanies the signal, when this is tuned in. With S2 closed, no modulation is present. The modulated


As can be seen here the oscillator is very small and is easy to construct.

signal is more easily found on a receiver or obtained through the IF amplifier, but the unmodulated signal is used for final alignment, in the way explained.

If an accurately calibrated receiver is available set this to $10 \cdot 7 \mathrm{MHz}$ and rotate the core of L1 to tune the generator accurately to this, with S2 closed for final adjustment. If an accurately calibrated signal generator can be borrowed (but will not be available later) set the generator to $10 \cdot 7 \mathrm{MHz}$, tune this in on another receiver, and adjust L1 (with S2 closed) for zero beat.

With an FM receiver with $10 \cdot 7 \mathrm{MHz}$ IF, place the alignment generator near this and adjust L1 to the correct frequency. If no such receiver is available, open S2 and place an insulated lead from the "RF" pin near the lst IFT of the IF amplifier, which should give a strong enough signal for the core of Ll to be adjusted. Subsequently, reduce the coupling as much as possible, so that the IFT cores can be peaked as described, with S2 closed. There is no need for the IFT's to be exactly on $10 \cdot 7 \mathrm{MHz}$ provided all the cores can be peaked and are still free to be moved either way. PART 1

AMPLIFICATION of d.c. and very low frequencies with very high gain, no valves or transistors, high output power capabilities-these are some of the advantages of magnetic amplifiers. Despite all these advantages, magnetic amplifiers are hardly ever used outside industrial electronics, yet they could be of great use in many amateur applications calling for low frequency amplification; applications such as motor speed control, photoelectric amplification; temperature control, or servo systems.

Probably the strongest reason for this is the lack of complete information on the working of a magnetic amplifier, textbooks either describing the amplifier as if magnetic theory were well known to the reader, or ignoring them altogether. This article sets out to fill the information gap.

## MAGNETIC CIRCUITS

One of the causes of the information gap is the way in which magnetism is taught in schools. We learn about magnetic poles and the simple "laws" of magnets as they were known 150 years ago, but nothing of the later developments in magnetic theory which made the solution of magnetic problems so much easier. We cannot, in this space, even attempt to outline the whole of modern magnetic theory, but we can take a look at the part of magnetism regarding magnetic materials and magnetic circuits which forms the basis of magnetic amplifiers.


Fig. 1 : Simple circuit to demonstrate Kirchoff's Second Law.

Figure 1 shows an electric circuit with a cell producing a current through two resistors in series. Now the voltage (electromotive force) across the cell is equal to the sum of the products of current times resistance for each resistance, or to use the mathematical form, $\mathrm{V}=\mathrm{R} 1 \times \mathrm{i}+\mathrm{R} 2 \times \mathrm{i}$, which is Kirchoff's Second Law. This electrical circuit is simple to deal with because the flow of current is exclusively in the circuit with no current flowing in the air around.

We can draw a very close comparison between the electric circuit of Fig. 1 and the electric magnetic circuit of Fig. 2, but with the exception that magnetic effects can be detected in the air around a magnet and do not confine themselves to the metal portion of the magnet. Designing magnetic circuits is rather like trying to design electric circuits to work in salt water; though the surroundings do not conduct so
well as the wires, they have a considerable effect, and current will flow even if there is a gap in a circuit.

If we imagine a magnet as a "ceIl", having two terminals and a magneto-motive force corresponding to the electromotive force of a cell, and the magnetic effect as a flow of magnetic current, or flux, then the similarity is obvious. Add to this the idea of magnetic resistance or reluctance for each piece of the magnetic circuit and we can make a Kirchoff's Law for the magnetic circuit.

$$
\mathrm{MMF}=\mathrm{R} 1 \times \mathrm{N} 1+\mathrm{R} 2 \times \mathrm{N} 2
$$

where $R$ is reluctance (magnetic resistance) and $N$ is magnetic flux.


Fig. 2: The magnetic equivalent circuit of Fig. 1.
This idea is useful in the design of transformer cores, relays, and other electromagnetic devices but cannot be used as simply as Kirchoff's Law, because the reluctance is not as constant as the simple resistance of Kirchoff's Second Law, but varies with the flux. Despite this complication, however, the idea of a magnetic circuit is useful, especially since the greatest reluctance in most magnetic circuits is the air-gap (or any other gap not closed by magnetic materials) which is constant at any value of magnetic flux.

It is rather like the case of a circuit of several resistances in series, all about $100 \Omega$ but varying with current (as thermistors do) except for one which is about $15 \mathrm{~K} \Omega$ and which "swamps" the others. In the same way, the large reluctance of an air-gap in a transformer core makes design easier because it makes it unnecessary to bother too much about the precise reluctance of other portions of the circuit.

Incidentally, the idea of MMF is used mainly with electromagnets. When permanent magnets are incorporated in a magnetic circuit, as in polarised relays, they are treated as a "negative reluctance" which contributes flux, rather than as a source of m.m.f.

## MAGNETIC FIELD STRENGTH AND FLUX DENSITY

If we wind a solenoid and pass a measured current through it, we find that the magnetic effect, provided that no iron or other magnetic material is present, depends only on the current $i$, and the number of turns of the coil per unit length. For a very long coil, the magnetic field strength is simply $\mathrm{H}=\mathrm{N} \times \mathrm{i}$, and H is measured in units of amperes/metre. As far as we are concerned this is a measure of the force which this solenoid would exert on a compass needle at a distance.

If a core of iron or any magnetic alloy is now placed inside the coil, the effect on a compass needle is greatly increased. The magnetic field has magnetised the iron, and the two different effects are adding together. We can measure this in the same way, but several old and confusing mistakes can be avoided if we measure this new effect by using induction. When a current changes in a coil which is wound on the same core as another, there is a voltage induced in both coils which is proportional to the number of turns and the rate at which current is changing. The quantity which changes the iron core is the magnetic flux which we met earlier and is measured in Webers or volt-seconds.

We find that the effect of a solenoid with an iron core can be measured by dividing the flux in the core by the area of cross-section of the core; this unit is called flux-density, and it is measured in Teslas ( $\mathrm{T}=1$ Weber $/ \mathrm{m}^{2}$ ) and given the symbol B .

## RELATION BETWEEN B AND H

If we set up the solenoid of the first example with an ammeter for measuring current, so that we can calculate H and a search coil close by connected to a flux meter so that we can measure $B$, we can place a magnetic material in the core of the solenoid and measure both $B$ and $H$ over a range of values. When we do so, we get a graph of the kind shown in Fig. 3. Starting from $\mathrm{H}=0$ and slowly increasing, the curve rises sharply and then flattens off until the rate of rise is equal to the rate of rise of $H$. When $H$ is decreased again, the value of $B$ does not take the same values as it did on the way up.


Fig. 3: Graph showing relation between ' $B$ ' and ' $H$ '. (Note; for 'remanace' read 'remanance'.)

The value of B left when $\mathrm{H}=\mathrm{O}$ again is called the remanance and measures the ability of the material to be permanently magnetised. To reduce $B$ to zero, an opposite value of H (called the coercivity) must be applied. As the values of $H$ in the opposite direction are increased, the curve traced out becomes the opposite of the one previously traced, but the first portion of the graph is never traced again unless the iron can be completely demagnetised (by heating, for example) and the experiment repeated.


Fig. 4: Simple magnetic amplifier. The DC input to the 'primary' is not a mistake! See text.

The interesting portions of this curve, called a hysteresis curve, for our purposes are the steep sides and the flattened top. When the current in the coil is at a value corresponding to point $A$ on the curve, the flux density changes considerably with a change of current, a condition which is ideal for a trans. former or choke since it means the greatest possible reverse voltage, and so the maximum inductance. When the current in the coil is at a value above point $B$, the change of flux density is very much less, and the material of the core is said to be saturated. In this condition, an increase of current in the coil causes very little increase of flux density, very small induced voltage, and so very little opposition to change of current (low inductance) so that large currents can flow. A core which includes an air gap limits the flux density because of the reluctance of the air gap and so reduces the possibility of saturation, which can have serious effects in transformers. A core which consists of a circle of magnetic material will saturate easily because of its low reluctance.

Many text books quote the permeability of a magnetic material $\mu$ as if it were a constant. As can be seen from the curve of Fig. 3, it is not a constant, but varies with the magnetic field strength, so that the only value which has any real meaning is the maximum permeability, the value of $\mu$ at point $A$ on our curve, because this indicates the usefulness of the material for a transformer core.

## A SIMPLE MAGNETIC AMPLIFIER

Figure 4 shows the circuit of the simplest form of magnetic amplifier. This consists of the choke L1 which is provided with two windings on a core of material which saturates easily (no air gap) and which has a very small area enclosed by the forward and return paths, meaning that the energy spent in magnetising and de-magnetising the core is not too large. Such a multiple winding choke (it is not a transformer, despite the similarity of the symbol) is termed a reactor. The winding marked 3,4 is used as a choke in an a.c. secondary circuit, and the winding marked 1, 2 is used to apply magnetic flux to the core by passing a steady current. The winding 1,2 is usually known as the control winding, and a

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choke La is wired in series with this winding to prevent a.c. being fed to the control voltage source through transformer action in the reactor.
Imagine that no d.c. is applied to the control winding. The winding 3,4 then acts as a choke (providing that the current through it is not too large) so that only a fraction of the alternating voltage applied to the secondary circuit can appear across the load resistor. If now some current flows in the control winding, the condition can be reached where the sum of flux due to the control winding and flux due to current in the secondary winding cause the core to saturate. At saturation, the inductance of the secondary winding is very low and a high current flows in the load, so that the average current through the load increases. This is equivalent to shifting the bias conditions of the amplifier, and this shift in the operating point is shown by the curves of Fig. 5.


Fig. 5: At $A_{8} \mu=\frac{B}{H}$ Is high, and the inductance of the coll (proportional to $\mu$ ) is high. At $B$ the reverse obtalns.

This type of magnetic amplifier is known as a straight-saturating amplifier and is seldom used because the sensitivity is low and the waveform across the load is distorted. It does, however, illustrate well the principles involved in magnetic amplifiers, and is easy to construct. The current amplification produced is equal to the ratio of turns of the control winding to the turns of the secondary winding (that is, the same as the ratio of a conventional transformer); note that the step up is from d.c. in the control winding to a.c. in the secondary winding so that this offers a method of controlling a.c. by means of a d.c. source.

## SELF SATURATING MAGNETIC AMPLIFIERS

For many purposes, greater sensitivity and a better waveshape are required and these requirements have resulted in the self saturating magnetic amplifier, which uses the secondary current to contribute more to the saturation of the core. This is achieved by making the secondary current travel only in one direction in the core. If the core is already saturated by the control current the inductance is very low and the secondary current is unimpeded.

## TELEVGIOR

 JULY ISSUE
## RENOVATING THE PHILIPS G6 CHASSIS

Continuing his series on renovating and fault finding in ex-rental colour chassis Caleb Bradley starts to work this month on the first Philips chassis. We've plenty of tricks to tell you on this one.

## DC RESTORATION CIRCUIT

Many single-standard monochrome chassis have a perfectly stable vision signal at the detector but then lose the black level through a.c. coupling in the video section. Keith Cummins has developed a circuit to give d.c. restoration at the c.r.t. cathode however. Development was undertaken using the popular BRC 1500 chassis but the circuit could be adapted for other chassis.

## WORKSHOP HINTS

Vivian Capel turns his attention to tuning capacitor gangs, both radio and TV receiver types, describing the fault conditions that arise and methods of dealing with them.

## COLOUR RECEIVER INSIGHT

Much insight Into colour can be obtained by examining the wide variety of circuitry used in different chassis. This month we take a look at the Korting 51763 series which incorporates many novel features including an unusual ident stage and tube protection on the event of fleld collapse.

## FAULT FINDING GUIDE

This month John Law swltches his attention to a typical l.f. panel (Bush TV141/Murphy V159 serles) and describes the faults that arise and trouble-shooting procedures to be adopted in this part of the receiver.

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## ELECTRONIC IGNITION

A recently announced transistorised ignition system from Dolphin Designs has its casing made from diecast aluminium with a special epoxy finish. All metal parts are cadmium plated and connections are through watertight glands. All metals are plated to prevent electrolytic corrosion and the circuit is built on a fibreglass board with a double thickness of copper. The unit is designed for 12 V cars with either positive or negative chassis.


A dash-mounted switch will revert the car to conventional ignition whilst driving either for comparison or the unlikely event of failure. The ignition system is priced at $£ 12 \cdot 95$, plus 30 p postage, plus VAT, and is available from: Dolphin Designs, 59 Leighton Avenue, Leigh-on-Sea, Essex.

## PEAK PROGRAMME METER

The PPM concept was developed by the BBC for checking modulation depths at transmitters, and the modern derivative (BS 4297) is used in the broadcasting and sound recording fields. The standards define attack and decay times and logarithmic meter scaling, with positive and negative peak detection as some sounds have + and - levels differing by up to 8 dB . The drive circuit, produced by Surrey Elec-

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Two other instruments in the Advance "pipeline" are the TC17 timer counter and DVM44 digital voltmeter. Advance Electronics Lid., Instrument Division, Raynham Road, Bishops Stortford, Herts. tronics is for 1 mA L.H. zero meters and is available as a kit at $£ 8$ or built and aligned at $£ 12$.
The attack time is 2.5 mS , decay rate 1 dB per 100 mS and detection + and - peaks within 0.5 dB . Display is a log. meter scale of 6 with 4 dB divisions ( -22 to +4 ) or BBC style 1 to 7 . Accuracy is within 0.3 dB and frequency response $-0.5 \mathrm{~dB}(20 \mathrm{~Hz}-$ $20 \mathrm{kHz})-1 \cdot 0 \mathrm{~dB}(12 \mathrm{~Hz}-30 \mathrm{kHz})$. Input impedance is $100 \mathrm{k} \Omega$ and supply is from +24 V to 50 V . Consumption is 25 mA .
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STANDARD portable transistorised equipment commonly operates from a 9 volt battery supply. When the occasion arises for such devices to be operated from a 12 volt supply, for example on a car battery, it is clearly necessary to provide some form of voltage dropper. This is not as simple as at first appears, due to the common use, for example, of class $B$ output stages in portable equipment. Since it is not a constant current load, therefore, it is impossible to insert a simple resistor to provide the reduced supply voltage.
Neither is a potentiometer arrangement adequate, since it requires a standing current greater than the drain of the device in use. Zener diodes face a similar objection, and the integrated circuits available to date have generally required complex interconnections and several discrete components to set the required output voltage. Hence the advantages of the TBA 435, a simple three terminal device specifically designed to provide a fixed output voltage of 8.5 volts when operated from any power source between 10.5 and 20 volts. Its applicability to the problem of operating 9 volt transistorised equipment from automobile power supplies is obvious.

## Performance

The device is capable of supplying continuously a current of over 100 mA . at the rated output voltage, representing an internal power dissipation of 0.75 watts, increasing to 4 watts if mounted on an


Fig. 1: Output regulation curve. Note action of protection circult as rated current is exceeded.
adequate heat sink. The circuit is, of course, fully protected against overload and even short-circuit conditions, while maintaining the rated output voltage to within $1 \%$ over the design range.

The operation of the protection equipment is fully illustrated in Fig. 1, which clearly displays the drop in output voltage if an attempt is made to exceed a safe current. When in automobile use, it may be found advantageous to precede the regulator with an RF choke, to prevent ignition interference entering the system through the power supply lines. However, the IC itself has a capacity to reject ripple and high frequency noise by a factor of some 50 dB , a property which is highly significant both for mobile and mains-driven equipment.


Fig. 2: Practical application of TBA435 and connections for 100 mA and 600 mA types.
It might be of interest to point out that the TBA 435 is merely one of a series of similar threeterminal integrated circuit voltage regulators with analogous operating parameters but with a range of output voltages. In any case where a power supply must provide a well regulated protected output at a fixed voltage, the appropriate unit from this series should certainly be evaluated for the application.

| Type | TBA | L | TBA | TBA | L | TBA | L |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 625A | 005 | 435 | $625 B$ | 036 | $625 C$ | 037 |  |
| Output voltage <br> Rated current mA | 5 | 5 | 8.5 | 12 | 12 | 15 | 15 |

Note: The NE555 reviewed in the May issue is made by Signetics and not SGS.

# pratitially Wireless commentary by IENTII 

NOT until one of my colleagues and regular readers reminded me of the passage of time was I aware that this contribution is the hundredth Old Henry has made to PW. Have you really endured me that long?

Anniversaries are a time for recall and for resolution. Recall is a facet of Henry's waffling that may at times have seemed to take precedence. Resolution should be


Brimming over with human kindness
saved for innovatory dates, the Chinese New Year, the day after the annual holiday, any Monday morning . .

Instead, if you'll bear with me, I would like to blast off about a couple of subjects that have recently roused my ire--and which, according to my critics, seem to exemplify Henry's approach to radio design, construction, and marketing.

First, the last. And I mean the retailer; your local friendly dealer, as the Radio \& Television Retailers' Association would have us believe. Brimming over with the milk of human kindness, intent on social justice and the welfare of his customers, ineffably altruistic . . .

Believe me, the radio retailer who really cares; who came up from bright emitters and the days of the catswhisker; who will spend his precious spare time finding better ways of providing P.A. for the nearby Old Folks Club; whose advice is geared to needs and not mere profit: that chap is as rare as a fridge in an Eskimo's igloo!

Monthly, one reads glowing tributes in the enthusiast magazines from readers who, despairing of help from local dealers, took their gear to the manufacturer, and had satisfaction. I feel a bit wry about that-better that the manufacturer should give his local agent good backing.

I have even read examples of 'turn-around' repairs being done on equipment for which I have had outstanding parts orders for months, and can only think, with typical Henryesque cynicism, that the manufacturer was really thinking of his 'public image'. But that's another story.

As good an example as any of the way 'the trade' does not care was to be seen at Sonex, where Donald Aldous wore himself into the ground in organising a series of lectures. During the two trade days, attendance at the lectures was abysmally small.

Practically the only interest was shown when a quadraphonic demonstration was given, when the lecture theatre became half-filled.

I don't want to enter the sur-round-sound arena just now, but would agree that discrete fourchannel is the only truly 'hi-fi' way of doing things: that, just as with gramophone records, mono/ stereo, f.m. broadcasting, commercial radio, colour TV and any other technical innovation, compromise has resulted from the needs of compatibility, and we shall probably have some inadequate system ultimately foisted upon us.

The slogan seems to be: 'Never mind what's best. What can we get away with?'

Very few dealers are today happy to take in repairs. They will tell you that their service department is already stretched to the limit to service the stuff they sell. which, when you think about it, argues that either their workshop is under-staffed or the stuff they sell continually goes wrong.

The truth is that there is little
profit in service. The retailer does not regard good service as necessarily fostering goodwill. In our own establishment, the 'Boss' is fond of saying: 'One second complaint outweighs twenty well-done jobs.'

Which brings me to my second grievance . . . . It is possible to make service pay, but only if it is run as a strictly economic business. And that, friends, entails our paying the true price for what we have done.

Take the case of the ordinary interconnecting lead. You can buy made-up packs of practically any plug or socket to any other for round about a pound, often less. But if you want a 'special' lead made by an otherwise busy engineer, you must expect to pay half as much again.

He has to fit this between other jobs, sort out matching more often than not, forage among the rubble for plugs and disentangle cable, set up his mini-vice and lick the end of his soldering iron, muttering incantations the while.

In a workshop like ours, where one of my colleagues waits till I am looking the other way, then 're-organises' all the drawers of

muttering incantations the while
small parts, it is lead-making that is most sick-making, and wildly uneconomic.

Perhaps that is why the small dealer no longer exists, and the package-seller who has taken his place shuns service: because we, you and $I$, are unwilling to pay our whack!


ENLARGER timers generally rely mainly on charging or discharging capacitors through variable resistors and an appropriate switching circuit. These timers can give a good repeat accuracy of $\pm 0 \cdot 25$ seconds but actual settings can only be as accurate as can be manually calibrated, using a stop watch, in most cases no better than $\pm 1_{2}$ second midrange.

This IC Timer relies for its accuracy on the mains supply frequency; it has a very high repeat accuracy, requires no calibration and is switchable from 0-9.9 seconds in 0.1 second steps and $0-99$ seconds in 1.0 second steps-the accuracy directly proportional to that of the mains. It has been my experience that the $0-9 \cdot 9$ second range is useful when printing in colour when filter changes dictate exposures of seconds and fractions of a second.

## CIRCUIT DESIGN

The circuit uses nine TTL i.c.'s. The timing is derived from 50 Hz mains supply frequency which is obtained from the secondary of Tl Fig. 1 and fed via a capacitor Cl to a two transistor Schmitt trigger circuit the output of which goes to a control gate. The output from this gate goes to an SN7490, IC4 in Fig. 2, used as a divide by five. The 10 Hz output from this goes to a divide by ten, IC5, to give 1 Hz . Two more SN7490's, IC6 and 7, are used as divide by ten circuits; the BCD outputs of each one being decoded by an SN7442. The ten outputs of the SN7442's pass to the single pole twelve way wafer switches S2 and 3.

To start the timing sequence Sl is depressed which causes a logic pulse to be fed to the input of a flipflop of an SN7472, IC3, one output of which goes to the control gate Gl so that 50 Hz is applied to the driving circuit; and to gate G2 which is connected to gate G8 controlling the driving circuit. The other output of IC3 goes to the zero reset of the dividers IC4-7.

If, for instance, S2 is on position 4 and S3 on
position 5 then when 45 seconds have elapsed the outputs from the wafer switches are at logic 0 instead of logic 1 as they would have been before the beginning of the timing sequence. These two outputs are inverted using two nand gates G3 and 4 the outputs of which are applied to another nand gate G5 whose output will be at logic 0 , when the 45 seconds have elapsed and at logic 1 at any other time.

The output is used to clear the flip-flop IC3 so causing it to revert to its original state which blocks the 50 Hz clock pulses, clears the dividers IC4-7 and
$\star$ components list



A Fig. 1: Circuit of 5V stabillised power supply and two transistor Schmitt trigger arrangement.

Fig. 2: Diagram of interconnectlons and supply leads for the nine

switches the triac off via the reed switch. Two nand gates are connected so as to eliminate contact bounce at S1 (gates G6 and 7).

Switch S4 is an added facility which allows the circuit to count to ten seconds in $0 \cdot 1$ second steps and this is achieved by merely short-circuiting one of the divide-by-ten i.c.'s IC5 causing 10 Hz to be applied to IC6 instead of 1 Hz . When either or both wafer switches are set on positions 11 and 12, closing S1 switches the triac on which is held on until Sl is closed again; this facility allows the enlarger to be on for focusing.


Fig. 3: Isolating stage using a reed relay to control the triac switch.


The IC board is mounted in the lid of the tox on spacers to minimise the length of the wiring to the switches.


Fig. 4: Placement and wiring of the IC's on the veroboard panel. Note carefully all breaks in copper rails.


Fig. 5: Layout of panel carrying the power supply, Schmitt trigger circuit and reed relay.

Although the output of the final i.c. (IC8) to the switching circuit is compatible with the gate current required for the triac, the triac cannot be driven directly as it would mean that all the i.c.'s would be live and could not be earthed. In order to overcome this problem a reed relay is driven from IC8 and the triac switched from that. Unfortunately the reed relay used is not TTL compatible so an interface circuit consisting of a 2 N 2926 is used between IC8 and the reed relay, Fig. 3.


The panel of Fig. 5 is mounted on insulating spacers on the aluminium partition as well as the 5 V regulator. The triac is mounted on a tag strip, part of the mains transformer.

Unfortunately, due to the high cost of i.c. holders (as much as twice the cost of some i.c.'s) they have not been used in the construction of this circuit. The i.c.'s have been mounted directly on to the $0 \cdot 1 \mathrm{in}$. matrix Veroboard, Fig. 4, care being taken not to allow the i.c.'s to get too hot although that is not much of a problem providing one solders them quickly and in rotation. The wiring between all components should be as short as possible (never any longer than 10 in .) so as to prevent spurious triggering of the i.c.'s.
The construction is fairly straightforward, the whole unit being mounted in an aluminium box. The switches S1-4 are mounted on the lid as is the IC board, held by 2 in . long 2BA nuts and bolts with spacers to provide a stand off so that the wires from S2 and S3 to the pins on the i.c. board are as short as possible. For convenience the triac was mounted on the transformer which was bolted to the case.

The power supply reed relay and interface circuit for the triac are mounted on Veroboard, Fig. 5, which is fixed to a piece of aluminium fixed across the inside of the box.

When the wiring has been completed and checked thoroughly (without the triac connected) it is possible to test the timing.

The switches S2 and S3 should be set to each of their positions and the timing tested: if it is not correct then one should look for faulty connections, missing drilled holes in Veroboard and solder bridges. When testing the timing it is important that the triac is not connected. Use a meter on the reed switch.

Because of the fact that the triac cannot be driven directly from the i.c.'s it is necessary to use an interface circuit. The i.c. board output goes to a resistor and transistor which switches hard-on when it receives a logic 1 voltage and a reed relay is energised which provides the gate current for the triac to turn on. Using a reed relay in this way the i.c.'s are completely isolated from the mains.

## INNEXTMONTH'S

## Practan Whilliss

Details of a new series of articles, for readers who want to get started in construction of simple radio and electronics projects. TO BE LAUNCHED

DON'T MISS ANY ISSUE OF PRACTICAL WIRELESS AUGUST ISSUE ON SALE JULY 6th

THe use of electronic devices to modify the sound of an electric guitar has become very popular in recent years, and the equipment of a guitarist often includes such devices as echo chambers, 'fuzz" boxes and 'wah-wah' pedals. This article describes a tremolo generator which can be used not only with electric guitars but also with most other electronic instruments.

## REQUIREMENTS

To add tremolo to a musical sound it is necessary to vary the loudness of the sound repeatedly at a very low frequency. In the terminology of electronics, this process is called amplitude modulation.

Figure 1 shows a block diagram of the tremolo generator. The input signal coming from the guitar has a more or less constant amplitude. The tremolo signal is generated by a low-frequency oscillator, whose output frequency and amplitude can be varied by means of foot pedals. The input and tremolo signals are fed into an amplitude modulator, whose output consists of the input signal with an amplitude which varies at the frequency of the tremolo signal.

Figure 2 shows oscilloscope traces of the signals involved. The input signal (a) is, in this case $a, 400 \mathrm{~Hz}$ wave generated by an audio signal generator. The tremolo signal (b) is a signal of much lower frequency, about 8 Hz . The output signal (c) consists of the 400 Hz input modulated at 8 Hz .

## CIRCUIT DESCRIPTION

The circuit diagram is shown in Fig. 3. It is easier to understand the circuit if the various sections are considered separately.


#  Gulffrs and <br> ORGANS 


R.MORGAN


Fig. 1. Block dlagram of the stages comprising the tremolo generator.

Fig. 2(a) left, shows an input slgnal of 400 Hz from a signal generator while 2(b), centre, is a much iower frequency signal from the oscillator In the tremolo generator. The two signals are fed to a modulator stage producing the signal shown in 2(c), right.


## Oscillator

The oscillator is required to produce a reasonably sinusoidal output at very low frequencies, from about 1 Hz to 10 Hz , and the usual circuit for this purpose is the Wien bridge oscillator. In Fig. 3 transistors Trl and $\operatorname{Tr} 2$ form a Wien bridge oscillator whose frequency is determined by resistors VR1+R3, VR2 + R4, and capacitors C2, C3. The frequency can be varied over the required range by varying VR1 and VR2, which are ganged.

This type of circuit requires an amplitude controlling device to keep the output amplitude at a level low enough to avoid saturation of the transistors; without this the output increases to a point where severe distortion occurs. The usual amplitude control device is a thermistor, which can be expensive, but by suitable modification of the circuit it is possible to use a much cheaper device, namely a 6V bulb LP1.
The output of the oscillator is fed into a potentiometer, VR4, the setting of which determines how much of the oscillator output is fed into the next stage. Consequently VR4 acts as a modulation depth control.

## Modulator

The musical signal from the guitar is fed into $\operatorname{Tr} 3$ and amplified. The degree of amplification depends on the current through $\operatorname{Tr} 3$ and if this current is varied the amplitude of the output from $\operatorname{Tr} 3$ is also varied. The current through $\operatorname{Tr} 3$ is controlled by $\operatorname{Tr} 4$, so that when the oscillator signal is fed into Tr 4 , the current through $\operatorname{Tr} 3$ varies at the same frequency as the oscillator signal. Consequently the amplitude of the output from $\operatorname{Tr} 3$ also varies at the frequency of the oscillator signal. The output from $\operatorname{Tr} 3$ is therefore the musical signal modulated at the frequency of the oscillator.



Fig. 3. Complete circuit diagram of the tremolo generator. Note that separate batteries are employed to avoid using expensive decoupling components.
$\star$ components list


## Output stage

The modulator output is at a high impedance and suitable for feeding directly into amplifiers with a high impedance input, i.e. all valve amplifiers and most transistor amplifiers. However, some transistor amplifiers have low-impedance inputs and if the modulator is used with such an amplifier an impedance-conversion stage is needed. The addition of such a stage makes the output suitable for all amplifiers, whether of high or low impedance, and there are certain other advantages such as a reduced level of mains hum. The output stage in the present design is an emitter-follower formed by Tr5.

## Power supply

If several parts of a circuit are fed from one power supply, there is a danger of unwanted interaction between one part of the circuit and another, due to voltage variations, unless the supply is de-coupled by means of a resistor and capacitor. When a circuit is designed to run at very low frequencies, as is the case here, the de-coupling capacitor has to be very large and therefore very expensive. A much cheaper solution is to use separate batteries for each part of the circuit. In the present design two batteries are used, one for the oscillator and the other for the rest of the circuit.

## CONSTRUCTION

The electronics are built on a piece of $0 \cdot 1$ in. matrix Veroboard, $11^{3}{ }_{8} \times 3^{3}{ }_{4} \mathrm{in}$. A suggested layout is shown in Fig. 4, but the layout is by no means critical and can be altered if necessary to suit components of different sizes and shapes. Constructors who are not used to building on $0 \cdot 1 \mathrm{lin}$. Veroboard should, take care to avoid bridging between the copper strips, by avoiding the use of too much solder.

The two sections of the circuit are almost completely separated electrically by severing all the copper strips where indicated, except for the negative rail, using a Vero spot cutter or a ${ }^{1}$ gin drill, or alternatively, with a sharp knife. The transistor leads are rather short and constructors who are inexperienced in soldering are advised to hold the leads with a pair of fine-nosed pliers during soldering, at a point between the solder joint and the body of the transistor.

Care should be taken to connect the input and output jack sockets correctly, with the negative rail to


Fig. 4 : Layout of the circuit board. Breaks in copper rails are best made before fitting components.
the socket contact which touches the outer conductor of the jack plug. If a mistake is made in this, an embarrassing amount of hum may be induced.
Potentiometers VR1 and VR2 are ganged linear controls. While ordinary carbon potentiometers are satisfactory, it is preferable to use wire-wound potentiometers because they have a much longer life and are less liable to noise. Ganged wire-wound controls are not easy to obtain, and for that reason the present design uses two ordinary single potentiometers with their spindles connected together. The mechanical details of this linkage using Meccano parts are shown in Fig. 5a. Because the two controls are "back to back", the connections may appear to be a little unusual, but if Fig. 5a is followed there should be no difficulty.
The mechanical linkages between the foot pedals and the controls are worth considering is some detail.


Fig.(5a), top, shows the assembly of the Meccano parts and potentiometers comprising the tremolo frequency pedal.
Fig. (5b), bottom, is the pedal assembly for the tremolo depth control.


Meccano collar with screw (No.59) soldered to brass angle (repeat on other side of pedal). After soldering drill through the brass angle to clear the axle.
Fig. 6: Details of the foot pedals which were eventually fitted to the prototype unit.

It is not possible to rotate one's foot through more than about $60^{\circ}$ without discomfort, while an ordinary potentiometer can rotate through about $300^{\circ}$. In the case of VR1 and VR2 a variation in resistance from zero to about $10 \mathrm{k} \Omega$ is required and this can be achieved simply by using one end of the track of a $50 \mathrm{k} \Omega$ potentiometer. In the case of VR4 it is necessary to use almost all the available rotation of the potentiometer so a 4 to 1 gear ratio is needed between the pedal and the control spindle. Meccano provides a simple and cheap way of building the linkage, and a suggested arrangement is shown in Fig. 5b.
From experience it was found that a sturdier pedal assembly was desirable and two were constructed, as in Fig. 6. They were each covered with corrugated rubber sheet.

A piece of equipment of this sort is subjected to a certain amount of rough treatment, and it is important to build a strong framework capable of withstanding abuse. A sug. gested design of framework is shown in Fig. 7.

The circuit board can be fixed with spacers to the back of the case. The on/off switch and the input and output sockets are fixed to the left-hand side panel. The two batteries are held in place by a metal bracket.

[^5]
## TESTING AND ADJUSTMENT

When the construction has been completed, and the circuit checked carefully, the unit is ready for testing and adjustment. It is useful to have an oscilloscope for this purpose, but it is not essential. Because of the use of two batteries, each part of the circuit can be tested separately.

## Oscillator

The first job is to set the position of the frequency pedal. Slacken off the grub-screws holding the short axle between VR1 and VR2 and turn the spindles of the potentiometers VR1 and VR2 to the minimum resistance position. Push the foot pedal to the extreme end of its movement, and tighten up the grub screws. Check that the potentiometers rotate smoothly when the pedal is moved. Leave the pedal near the middle of its travel.

The next adjustment is to set the position of the modulation depth pedal. Slacken off the grub-screw of the cogwheel which drives the spindle of VR4. Rotate the spindle to bring the slider to the earthed end of the track. Push the foot pedal to the extreme end of its movement, and tighten up the grub screw. Check the rotation of the potentiometer when the pedal is moved and once again leave the pedal near the middle of its travel.

Connect the oscillator battery (not the modulator ends of VR4. Set the timebase of the oscilloscope to a slow scan (about one trace per second if possible) and the input sensitivity to about 1 volt for full deflection. If an oscilloscope is not available, put the unit in a partially darkened room in a position where the lamp LP1 can be seen easily.

Connect the oscillator battery (not the modulator battery) and switch on. Adjust the preset potentiometer VR3 until the circuit starts oscillating, as evidenced by the trace on the oscilloscope or by a dull pulsating glow in the lamp LP1. Move the frequency pedal to bring VR1 and VR2 to a highresistance position. The frequency should drop to about 1 Hz or less. If the circuit stops oscillating,



Fig. 7 : Details of the casework housing the tremolo generator.
re-adjust VR3. Gradually raise the frequency by means of the frequency pedal and check that the oscillator works correctly over the required range of frequency; if it shows any sign of stopping, readjust VR3. The ideal position for VR3 is one in which the oscillation is just able to continue over the whole range of frequency, but the setting is not very critical. Finally, switch off.

## Modulator

Disconnect the oscillator battery and connect the modulator battery. Plug the guitar lead into the input socket and plug the output into the guitar amplifier. Switch on and check that the volume is satisfactory and that there is no distortion of the sound. If all is well, no further adjustment is needed. If there is noticeable distortion, it is probably due to overloading of the modulator, and this can most
easily be prevented by connecting a resistor of a few kilohms in series with the input, between the input jack and C5.

## OPERATION

Connect both batteries, the guitar and the amplifier. Switch on and adjust the positions of the pedals while playing long chords. The effect of the tremolo generator on the sound should now be apparent.

The guitarist may like to try the tremolo effect in conjunction with other devices. It appears to be particularly effective when used with a 'fúzz' box. It seems that better results are obtained when the 'fuzz' box is placed between the guitar and the tremolo generator, rather than between the tremolo generator and the amplifier, but this is a matter for experiment.


## 3-BAND TRANSMITTER PART 2 F.G.RAYER G30GR

THOUGH this VFO is particularly intended for use with the transmitter described last month it can be employed for the frequency control of any similar transmitter. The VFO has a straightforward circuit which can give reliable and excellent results.

## CIRCUIT

Fig. 1 is the circuit which uses a 6C4 as a Clapp oscillator, the HT to this stage being regulated by the OA2 voltage regulator. The valve capacitances are shunted by the large value capacitors C2, C3, and C4, so this type of circuit proves to be stable and trouble-free.

Components in the VFO stage are enclosed in a metal box on the chassis and this complete screening avoids any difficulties from stray feedback. The whole VFO is in turn enclosed in a metal box or case.

A 6BH6, V2, is a buffer-amplifier which also frequency doubles for working on 7 MHz and 14 MHz . On $3 \cdot 5 \mathrm{MHz}$ resistor R 6 forms the anode load. Output to the transmitter is by a co-axial lead plugged into the socket which is provided on the transmitter. The latter also supplies power for the VFO. When working on 14 MHz , the crystal oscillator stage of the transmitter acts as a second doubler.

Coil L1 and capacitor VCl tune $3500 \cdot 3800 \mathrm{kHz}$ and output frequencies are fundamental or doubled which makes the VFO calibration easy.


Fig. 1: Circuit of the two valve variable frequency oscillator. The power for the VFO is taken from the main transmitter, described last month. Components associated with the 6C4 oscillator valve are shown within the dolted enclosure.

## VFO BOX

The variable frequency oscillator coil LI and associated components are completely enclosed in a box, Fig. 2 which is very easily made from "universal chassis" members. A flanged member $8 \times 3 i n$. is taken and a V-shaped notch is cut in each flange 2 in . from the end. A right-angled bend can now be made $2 i n$. from each end. A sharp bend is best obtained by holding the member firmly on a piece of wood just large enough to rest between the flanges, and with one edge level with the bending line. This gives a U-shaped piece 2 in . high and 4 in . wide. The front is closed by attaching a $4 \times 2 \mathrm{in}$. flanged plate with four bolts. After all assembly is finished, a flat back plate $4 \times 2 \mathrm{in}$. is fixed on with self-tapping screws.

Tuning capacitor VCl is mounted on the front flanged plate as in Fig. 2 and the ball drive is put on. Its lug is held with a long bolt with spacer, or extra nuts.

Trimmer TC1 is fixed on a small bracket under a hole, so that it can be adjusted from above with a screwdriver. Tag-strips are used to anchor R2, C5 and RFCl. The coil L1 is positioned as shown, with its core adjusting screw projecting from the top of the box.

Wiring and components in the box should all be rigid.


Fig. 2 : Layout and wiring of components within the separate osclllator box.

Coil L1 is wound on a cored former, ${ }^{1}$ in. in diameter and $1_{4} \mathrm{in}$. long, with 30 turns of 26 swg enamelled wire, closewound. The winding is started as near to the top or pin end of the former as possible. The former can be lightly smeared with a suitable adhesive (Denfix) before winding. The end is secured with cotton and adhesive.

The VFO can be tested by applying heater and HT voltages (about 150 V ) and adjusting TC1 and Li for suitable coverage. With VC1 almost fully closed, L1 core is rotated to obtain $3 \cdot 5 \mathrm{MHz}$. VCl is then set almost fully open, and TCl is adjusted for $3 \cdot 8 \mathrm{MHz}$. One adjustment modifies the other, so this needs to be repeated several times. Sufficient coupling to a receiver is obtained by placing the aerial lead near the VFO.

Final calibration should be with the aid of a 100 kHz crystal marker or equally accurate frequency

standard. This will give calibration marks at $3 \cdot 5$, $3 \cdot 6,3 \cdot 7$ and $3 \cdot 8 \mathrm{MHz}$. Tuning the receiver to the second harmonic of the VFO will give $3 \cdot 55 \mathrm{MHz}$ (at $7 \cdot 1 \mathrm{MHz}$ ) and so on, while the fourth harmonic will give calibration for the 20 m band. When the VFO is being calibrated, or is in use with a transmitter, the higher frequency bands are direct multiples of the $3 \cdot 5 \mathrm{MHz}$ band frequencies, so working with and marking the scales is very straightforward.

## CHASSIS AND CASE

The main chassis is a universal chassis flanged member $8 \times 4 i n$. This allows similar members to be used for the top and bottom of the VFO case, with $7 \times 4 i n$. flanged members for the sides. The front is a flat plate $8 \times 7 \mathrm{in}$.


Rear view of VFO shows position of oscillator stage on maln chassis.


The long flanges of the chassis member are first cut away at each corner just sufficiently to allow this item to fit inside the flanges of the members which form the sides of the case. The chassis is bolted so that its top surface is 2 in . above the bottom edge of the $8 \times 7 \mathrm{in}$. panel plate. Voltage regulator VR and V2 are placed as in Fig. 3, and the wired VFO box is bolted in this position and its back closed with the $4 \times 2 \mathrm{in}$. plate.

## WIRING

Fig. 4 shows components and wiring under the chassis, with leads passing through a hole for heater, regulated HT and the connection from C5 to the grid of V2.

A tag strip holds flexible leads for $6 \cdot 3 \mathrm{~V}, 250 \mathrm{~V}$ and common chassis return. These three leads can be about 18 in . long and they are soldered to the pins of the 3 -way plug, which fits the transmitter power outlet. Power switching of the VFO is controlled by the CW/AM/OFF/NET switch on the transmitter.

The VFO output lead is about 9 in . of low-loss coaxial cable, passing through a grommet in the front panel and anchored at a tag strip. The VFO output
position is adjacent to the transmitter VFO input socket, so a long co-axial lead is not required or desirable. A long lead here may cause difficulty in resonating L2, or in securing adequate drive, due to its capacitance and possibly losses in the high impedance anode and grid circuits.

L2 has 38 turns of 32 swg enamelled wire closewound on a ${ }^{1}$ in. diameter former with adjustable core. The winding should begin as near to the tagged end of the former as possible.

The VFO should be tested before completing the case. It was found stable when used without the case, but the latter improves screening and rigidity. When the case is assembled, a short screwdriver is necessary to reach L1, L2 and TC1.
To assemble the case, put the side members with their flanges in the slots provided. The bottom and top can then be put into place, and are bolted to the sides. The case is checked for squareness, and the panel is correctly positioned. Holes for self-tapping screws are then drilled through the panel and side flanges, so that the panel can be secured. It is best to drill both panel and flange with a bit suitable for the self-tapping screws to hold, then use a larger bit, on the panel holes only for clearance.

The flanges of the member forming the chassis are bolted to the sides of the case.

## USING THE VFO

For operation on 80 m , VFO, driver and PA switches are all set to the 80 m position. With the transmitter control switch at "Net" grid current should be shown by the transmitter meter and the carrier should be


Fig. 5: Modified oscillator circuit to improve bandspread coverage on the higher frequency bands.
found on a receiver at the VFO frequency. The transmitter driver stage screen grid potentiometer will need to be turned back considerably to keep grid current down to the correct level.

For working on 40 m , place the VFO switch at $40 / 20 \mathrm{~m}$ and the transmitter switches at 40 m . L2 may need adjusting as mentioned for 20 m .

When first testing the VFO on 20 m , tune to about $14 \cdot 2 \mathrm{MHz}$. Set the VFO switch at $40 / 20 \mathrm{~m}$ and both transmitter switches at 20 m . With the control switch at "Net" rotate the core of L2 for maximum grid current as shown by the transmitter meter. On all bands tune the PA grid for maximum grid current, adjusting this to a suitable level with the drive potentiometer.

## CALIBRATION

The scales are marked so that the higher frequency bands are direct multiples of the low frequency band. So $3 \cdot 5 \mathrm{MHz}$ is also $7 \cdot 0 \mathrm{MHz}$ and $14 \cdot 0 \mathrm{MHz}$. As mentioned this greatly simplifies calibration.


Fig. 6: Suggested circuit to give full scale coverage on all bands, 1.8 - 28 MHz .

With this method of covering higher frequency bands only a smaller part of the tuning spread is used for 40 m and 20 m . During experimental use of
the VFO, two methods were employed to spread the tuning on the HF bands. These may be of interest, though they were not used in the completed VFO.

The method shown in Fig. 5 gives coverage of $3 \cdot 5-3 \cdot 8 \mathrm{MHz}$ on one range for 80 m . On the other range the VFO tunes from $3 \cdot 5 \cdot 3 \cdot 6 \mathrm{MHz}$, giving coverage of $7 \cdot 0 \cdot 7 \cdot 2 \mathrm{MHz}$ and $14 \cdot 0 \cdot 14 \cdot 4 \mathrm{MHz}$ with almost the full 180 -degrees rotation of the tuning capacitor.

With the switch at $80 \mathrm{~m}, \mathrm{VC1}, \mathrm{VC} 2$ and VC3 are all in parallel, forming a 75 pF variable tuning capacitor, with TCl and the 200pF capacitor as before.

When the switch is on $40 / 20$, TC2 with its parallel fixed capacitor substitute for VC2 and VC3 and coverage is restricted to that of VCl , that is $3 \cdot 5$ $3 \cdot 6 \mathrm{MHz}$, giving actual ranges of $7 \cdot 0 \cdot 7 \cdot 2 \mathrm{MHz}$ and $14 \cdot 0-14 \cdot 4 \mathrm{MHz}$.


A closer look at the VFO dial calibration.
A Jackson miniature 25 pF 3-gang capacitor was used. TC2 is adjusted so that $3 \cdot 5 \mathrm{MHz}$ falls on the same tuning position, with the switch at 80 m or $40 / 20 \mathrm{~m}$. The 80 m range has to be individually calibrated. The 20 m range is the second harmonic of the 40 m range.

Fig. 6 shows a modification used for some considerable time in conjunction with an all-band transmitter. With the switch placing L1 in circuit, TCl is adjusted for coverage of $1 \cdot 75-2 \cdot 0 \mathrm{MHz}$. This is for Top Band ( 160 m ) and 80 m , the $1 \cdot 75 \cdot 1 \cdot 9 \mathrm{MHz}$ portion of the band being doubled for $3 \cdot 5 \cdot 3 \cdot 8 \mathrm{MHz}$. L1 is 60 turns of 36 swg enamelled wire, closewound on a ${ }^{1} 2 \mathrm{in}$. diameter cored former.

When the switch places L2 in circuit, coverage is from $7 \cdot 0-7 \cdot 2 \mathrm{MHz}$. This gives operation at the fundamental frequency of $7 \cdot 0-7 \cdot 1 \mathrm{MHz}$ for 40 m . For $20 \mathrm{~m}, \quad 7 \cdot 0-7 \cdot 2 \mathrm{MHz}$ covers $14 \cdot 0-14 \cdot 4 \mathrm{MHz}$, with doubling. This same VFO band provides $21 \cdot 0$ $21 \cdot 4 \mathrm{MHz}$ for 15 m and $28: 0-28 \cdot 8 \mathrm{MHz}$ for 10 m .

Where the transmitter has few frequency multiplying stages this arrangement is very useful indeed. When the transmitter covers all bands with the minimum of circuits, it is helpful to fit a 4-way switch for V2 in the VFO. In addition to the untuned load, coils for $3 \cdot 5 \mathrm{MHz}, 7 \mathrm{MHz}$ and 14 MHz may then be selected, giving a choice of VFO outputs, with only tripling required to reach 21 MHz , or doubling for 28 MHz .


Following our review in the April issue of the extensive collection of vintage wireless equipment belonging to George Weston of Auckland, New Zealand, we received a copy of a letter sent to Mr. Weston by Mr. W. W. Barnes, G2FI, of Wadhurst in Sussex, who began his career in wireless in 1917 when he was with the Royal Flying Corp. He rightly congratulates Mr. Weston for having stuck to his task of gathering together so much ancient gear in a part of the world where the amount of such equipment must be very limited.

Mr. Barnes joined the Philips Company in 1926 and stayed with them for many years, mainly in their Servicing Department, and thus was able to offer some additional information on the 2510 and 2515 receivers that we pictured in the April issue. These sets were made at Eindhoven in Holland but when sent to England were fitted with Mullard valves and the type numbers changed to 2511 and 2516 respectively! Needless confusion does not seem to be the prerogative of the modern society! The 2511 was the first domestic receiver having two HF stages to appear on the English market and, according to Mr. Barnes, was far ahead if its time. It was built 'like a battleship' and reputedly was just as heavy! Obviously the forerunner to the AR88!

Mr . Barnes recalls feeding the output of a 2511 into a pair of PX4's which fed a mains energised 18in. loudspeaker. The speaker and its power supply alone weighed 48lbs.! We understand that the current copy of the Philips News Sheet refers to a letter from a customer saying his 2515 is still in use and functioning perfectly. Mr. Barnes built his own two-HF-stage receiver in 1922 from a design published in the old 'Modern Wireless'. However it was unstable to the extent that the coils had to be wound with resistance wire to 'hold it down'.

## Uransatlantic 巩 $\mathcal{X}$

Another old-timer from the ranks of the amateur fraternity is George Tagg, G8IX, now of Stoke-onTrent, who has been good enough to trust us with the original negatives and photographs of the wireless gear that he operated over fifty years ago, in 1919/20 to be precise. He was then a Leading Telegraphist at HM Wireless Telegraphy station at Mount Pearl, St. John's, Newfoundland and the equipment in use included a Model L receiver and a Poulsen Arc transmitter. (If anyone has any information on this receiver we would be glad to hear from them.)


George Tagg seen here at St. John's, Newfoundland, is still active on the air as G81X.

Remarkably enough the best DX included contacts with the station at Pembroke Dock. George recounts taking photographs of Alcock and Brown before their epoch-making Transatlantic flight and also making special transmissions and test signals for them while they prepared their equipment.

## $\mathbb{C}$ arly Courtesies

Our collection of photographs of 'very ancient stations' has been swelled by a contribution from Mr. S. D. Mason of Torquay whose station was active in 1910! It doesn't take much work on the office slide rule to realise that Mr . Mason is now a proud 80 years of age. In those early days he was located at Forest Gate in London and his call sign was just a simple 5SV. The preamble to his licence ran 'To all whom these presents shall come, the Post Master General the Rt. Hon. Herbert Samuel sends greetings... etc.' .. . obviously a unique document one would be very happy to possess.
Morse and telephony were both used at 5SV to work other stations in the Ilford and Barking areas but seemingly the ship's operators in the Royal Albert Docks were not very favourably impressed by the virtually untuned spark gap at 5SV! The aerial was a twin wire inverted L but only 30ft. long which must have been very short compared with the wavelength in use at the time.

At the beginning of the First World War all Mr.

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The spark gap at 5SV can be seen on the top shelf under the copper strip hellx coll which 'tuned' the transmilter. Below is the 2-valve transmitter consisting of an osclliator and modulator.

Mason's equipment was impounded and he found himself with the Royal Naval Air Service at the White City in London, very close to the present site of the BBC TV service. After operating WT equipment from airships based at near-by Wormwood Scrubbs (another famous name!) 5SV passed to Manston for flying training and finished up with No. 3 Wing in France. Unfortunately Mr. Mason is no longer on the air, which is rather unusual. Generally the bug, having bitten, doesn't let go! At least not for long!


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## SHORT WAVE DX

by MALCOLM CONNAH

APART from the gems of information which can be gleaned from this monthly page, the best sources of DX information and station news are the DX programmes broadcast by many stations throughout the World.

The following list shows the times at which some of the more popular programmes are broadcast. In all cases the broadcasts are in English.

Australia "DXers Calling" is broadcast during the English transmissions on Saturday at 2130; Sunday at 0430, 0730, 1300 and 1530 and on Monday at 0215.

Ecuador HCJB in Quito broadcasts its "DX Party Line" programme on Monday at 0030, Tuesday at 0300 and 1930 and on Wednesday at 0900.

Great Britain The BBC's "World Radio Club" programme is broadcast over the normal World Service outlets at the following times: Sunday at 0815, Wednesday at 1330, Thursday at 2030 and Friday at 2345.

Portugal Adventist World Radio transmit a programme of DX News compiled by our own World DX Club at 0935 on Sundays on 9670 kHz .

Sweden The "Sweden Calling DXers" programme is 25 years old this year but can be referred to as the "grandaddy" of DX programmes. The programme is broadcast every Tuesday and takes up half an hour of the normal English programmes.

Switzerland "Swiss Shortwave Merry-go-Round" is broadcast in all the English transmissions of the Swiss B.C. on the 2nd and 4th Saturdays of the month.

## Readers' Logs

The first item in this section this month comes from John Godwin of Rugeley in Staffordshire. John reports hearing Radio New Zealand between 0600 and $0800^{\prime}$ on 11780 kHz on his pre-war HRO receiver and "Joystick" antenna.

The transmitter power of Radio New Zealand is only ${ }^{1}{ }_{2} \mathrm{~kW}$ and reception in this country is fairly difficult. Well done John!

Alexander Morton of Framlingham in Suffolk has two receivers, a Bush AC2, 5 valve domestic and a VEF-204 10 transistor portable. His log included:
5920 R. Kiev, in English at 1930.

9023 Voice of Iran in English at 2000.
9690 R. Pakistan with news in English at 1815.
9745 HCJB, Quito, Ecuador, DX Partyline at 0900.
10040 Voice of Vietnam, Hanoi, news in English.
15415 R. Kuwait in English at 1630.
17895 R. Portugal in English at 1345.
Ian Gordon of Rubery, Birmingham has managed to extract signals from the following stations out of the central heating radiator which he use as an aerial and coax them into his Codar CR70A receiver:
9005 R. Iran in English at 2015.
9570 R. Australia at 0730.
9655 R. Damascus, Syria in English at 2050.
9670 Adventist World Radio at 0940.
9745 R. Baghdad, Iraq in English at 1950.
11675 R. Pakistan in English at 2035.
11710 RAE, Argentina, s/on at 2300.
11765 R. Australia, DXers Calling (Sun.), 0730.
15012 Voice of Vietnam, Hanoi in English at 1815.
17885 R. Japan, s/on at 0800 in English.
Shaun O'Sullivan of Keynsham (yes that's K-E-Y-N-S-H-A-M of Radio Luxemburg fame), Bristol, has a one valve 'H.A.C.' t.r.f. receiver and a two transistor amplifier connected to a 100 foot long-wire antenna. This combination enabled him to hear:
5930 Radio Prague, Czechoslovakia in English at 2135.

5995 Voice of America in English at 2040.
6160 Radio Sofia, Bulgaria in English at 2135.
6190 Vatican Radio in English at 2100.
7090 Radio Tirana, Albania in English at 2040.
7100 Radio Moscow, USSR at 2110.
Dave Gregory of Plymouth in Devon has a Bendix RA-1 aircraft receiver and an 80 foot long-wire antenna, his $\log$ for the month included:
7100 R. Pakistan in English at 2200.
9125 Test transmission from Panama City at 0035.
9535 Swiss B.C., Berne in English at 1530.
9725 Kol Israel noted at 2140.
9770 R. Baghdad, Iraq at 2045.
11640 R. Bangladesh in English at 1800.
11755 AIR, Delhi in English at 2230.
11805 Radio Finland noted at 1910.
11880 Radio Sweden in English at 1230.
11900 RSA, South Africa in English at 1115.
15175 Radio Norway noted at 2200.

## DX Clubs

The Youth DX Club International (YDXCI) which was mentioned in this column last month has announced that the maximum age of its members will be 17 years. The membership fee is $£ 1 \cdot 20$ which includes the cost of sending the magazine to any country. The address for membership enquiries is YDXCI, Bartletts, Culmstock, Cullompton, Devon.

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MEDIUM WAVE BROADCASTS by CHARLES MOLLOY

BILL JARVIS (Luton) has received direct from the BBC " 20 BBC Local Radio Stations" which gives details of locations and power.
R. Howard (East Runton) has sent the following log of BBC Local Radio Stations heard at his QTH in Norfolk-Radio Blackburn on 854 kHz ; Radio Sheffield 1034 kHz ; Leeds 1106 kHz ; Newcastle-onTyne 1457 kHz ; Teeside (Stockton) 1594 kHz . He also reports hearing La Coruna, Spain on 638 kHz ; AFN Bremerhaven 1142 kHz and Radio Sud, Andorra on 818 kHz . Andorra is a small independent state situated in the Pyrenees between France and Spain. Its other medium wave outlet is Radio Andorra on 701 kHz and both stations will verify a correct reception report written in English, with a colourful QSL card. It appears that Andorra no longer broadcasts on the short waves and it is only on the medium waves that the DXers can hear broadcasts from this mini-republic. Reception reports should be sent to Radio Andorra, B.P.I, 66720 Andorra-la-Vieille,

Andorra for the transmissions on 701 kHz and to Sud Radio, B.P. 7 Andorra-la-Vieille, Andorra for those on 818 kHz . Remember to include an International Reply Coupon (obtainable from main Post Offices) to cover return postage and to ensure a reply.
D. O. French (Stroud Green, London) is unable to hear the religious broadcasts from Trans World Radio Monte Carlo on 1466 kHz because of interference from the BBC Radio London station on 1457 kHz which transmits from Brookmans Park. He expects TWR to increase power on 1466 kHz from 400 kW . to 1000 kW by the end of 1974 . TWR will then join the growing list of medium wave broadcasters that use powers of a megawatt or more. Others include Radio Bucharest, Rumania on 1052 kHz ; El Beida, Libya 1124 kHz and Tripoli, Libya 1250 kHz both of which are easily heard in the UK after dark; Batra, Egypt 1106kHz; Skopje, Yugoslavia 809 kHz ; West Germany on 1421 kHz and Radio Luxemburg 1439 kHz . Other more distant stations that have been heard in the UK include Riyadh, Saudi Arabia on approx. 588 kHz ; Quazvin, Iran 841 kHz (now with 2000 kW ); Rajkot, India on 1070 kHz ; Urumchi, China 1525 kHz (usually in Russian with jamming) and the Voice of America Thailand 1580 kHz in English at 2200 hrs GMT. There is also the 1000 kW outlet at Calcutta, India on 1130 kHz which has been heard by the writer during July at 2200 hrs GMT with sign-on at the start of its Chinese programme. It is worth looking for this station between the burble on 1133 kHz and Barcelona/Libya on 1124 kHz .

## VHF/FM DXing

## by SIMON DAVID

FIRST of all, my thanks to all those readers who wrote to me; I shall try to pick out interesting points from as many letters as my modest half-page allowance will permit and combine these with any news items that are useful.

Many of you take Band 2 seriously as DXers and this is really the basis on which this feature was started, but there are times when weather conditions, particularly in the summer months, can produce some odd twitterings even on the usual entertainment programmes from local stations.

Mr. C. Rahn of Chesterfield has made a study of reception conditions and confirms that humid overcast atmospheres and very warm temperatures bring in some long distant signals. Of course, it is of little help if conditions are cooler en route, so when you have one of those sultry sticky days and the weatherman says that this is spread over the whole country, forget the lawn mower and switch on your f.m. receiver.

It does help to have a multi-element array and if you can arrange for it to be rotated easily, then you are in for a mixed bag of DXing. Mr. Rahn has four aerials: a roof ' H ' set to Sutton Coldfield; an indoor 'H' rotatable; a piece of unscreened wire about lft. long; and an indoor telescopic. The most fruitful, of course, is the rotatable which brings in most of the BBC local radio stations from afar, plus numerous foreign stations. However, even Caen $95 \cdot 6 \mathrm{MHz}$ was picked up by the telescopic. His tuner is the Omega ST-300C with a stereo decoder A1005M (G. W. Smith). Now with his new "weapon", a 6 -element rotatable, he is getting even better results.

There were disturbances on April 1st due to an aurora borealis. Although not visible due to overcast skies, the effects on v.h.f. broadcasting were a
"burbling" note on continental stations, detected by Mr. R. Ham in Sussex on 42 and 72 MHz . These noises also interfered with television on channels 3, 4 and 5.

The target for most DXers is to see how far into Europe one can penetrate on Band 2. Mr. G. P. Stanbury tells a fascinating tale involving an Antiference 4-element with two extra directors, a Rigonda "Symphony" radiogram with home made tuning meter in place of the magic eye, the f.e.t. preamplifier from April's P.W., and a rotatable aerial mounted on a cable controlled pulley system. In spite of some backlash, he has heard at his home in Chelmsford, Essex, many BBC stations as far away as North Hessary Tor and Wenvoe.

Continental pick-ups include five Dutch and three Belgian stations on Dutch network RTB-BRT, the latter being Aalter on $90 \cdot 4,95 \cdot 7,98 \cdot 6 \mathrm{MHz}$ and Veltem on 89.5 and $93 \cdot 7 \mathrm{MHz}$. West German stations received include Langenberg and BFBS Cologne. There are local disturbances including a 400 ft . hill and a police transmitter half a mile away that almost swamps the continentals at times. There is, he says, a local river board authority transmitter which appears to be "ill-matched to its damp string!"

One of the problems of DXing in residential areas is the interfering reflections from neighbouring aerials, and badly aligned tuners without i.f. traps on the aerial can cause trouble. Mr. Stanbury appeals to all users to fit i.f. traps to keep the i.f. on a leash and I would endorse this wholeheartedly.

In the May article I mentioned rejection filters to suppress adjacent channels; the Wolsey Model No 2 covers 40 to 100 MHz range with maximum rejection of 45 dB . Most good aerial dealers should have these for around a fiver. Finally, the IBA have announced on their medium wave test transmissions on 557 kHz that v.h.f. tests start in the autumn on $95 \cdot 8 \mathrm{MHz}$ (entertainment), $97 \cdot 3 \mathrm{MHz}$ (news).


AMAHFU: BANDS

## SHORT WAVES

## by DAVID GIBSON, G3JDG

Agood month for the persistent listener and some good DX loose on all bands-a fair assessment of the Amateur bands these past few weeks. Eighty metres has yielded such little treasures as ZL2BT and DU1EJ, and even 7 MHz has seen fit to part with items like VK5CX and 5R8BD. Good DX on twenty and fifteen but the surprise came from 28 MHz which has had spasmodic bursts of life. When this happened the DX was usually 599+ or $5 \& 9+$. Goodies on 28 MHz logged include TU2DD, VP2LI and ZMIHV.
Occasionally one hears of an Amateur station accused of causing interference (QRM) to other receiving equipment, usually television. If this happens, the Amateur involved will take steps to solve the problem and help effect a cure for such QRM.
The boot is now on the other foot. There seems to be a marked increase in the number of intruders into the Amateur bands. Not only this but worse; the interference caused by these illegal intruders is deliberate. Other Hams have reported this also.
John Bagley (Birmingham) comments that we should find out where the QRM is originating from and then refuse to work Amateurs in that country (ies). It might, perhaps, be better to do exactly the reverse-make a point of working these localities and then diplomatically raising the question openly on the air i.e., "Do you suffer from QRM OM there seems to be a strong source somewhere in your area?". The RSGB has set up an Intruder Watch. Perhaps its findings could be published?
A couple of hot tips have arrived. Listen 1425014300 kHz at 1230 hrs . at weekends for YAlAH and YA0CDR. There is a net on Friday mornings ( 14250 -14300 kHz ) called the Arabian Nights (Knights) Net. Stations from VK and ZL frequently call in.
Apparently that XG prefix heard by many mystified souls earlier this year was used by XE1J. Something to do with a 450th anniversary. Another callsign was changed because of the seventh Games held in Panama City. When will $G$ stations get in on this callsign changing act? There are many events in England which merit an instant ohange of callsign:
hearing the first hedgehog; Noddy's birthday.
In the April issue I asked the real G4GJ to "stand up". Well, he has! Operational from Bingley in Yorkshire since 1945 this station has a daily sked (not weekends) with G3CJS on 3680 kHz . These stations have been doing some experiments with loop aerials.
Lionel Outridge (Canterbury) read my remarks re 4 metres ( 70 MHz ), and promptly built a converter using dual-gate MOSFET's (r.f. and mixer) plus a MPF102 crystal oscillator. Lionel says that a nearby BBC relay on channel 5 puts out a string of 50 Hz modulated carriers right across the $70 \cdot 1-70 \cdot 3 \mathrm{MHz}$ band which makes reception difficult. (Who do you report the BBC to?). Other equipment is a FRDX400 and an indoor dipole. Stations received on 70 MHz ; G3-AJS, FDW/P, FIJ, FZL, IIR, JEQ/P, LVK, LVP, LVP/M, LVP/P, OHH, OQT, ORH, ORH/M RUA/P, TDM/P, VER/P, VHH, VPK, VPS/P, WMR/P, YCN, YTM, ŹFP, ZRH, plus GB3SU, G4BLN, GW4ABR/P, GW4BGG/P and G6SG. So the band isn't morgue. How about a few more of us breathing life into it?

Anyone hear 5GAA, HA25KFZ or O1KA? Gareth George (Cheltenham) did and asks what, whom, who or where? (Perhaps someone else heard the first hedgehog?). Receiver is an HRO with a 100ft. long wire. A twiddle on twenty brought: EA3GN, EA5FQ, EA7QG, IT9JC, JA6BEE, JA6NPF, JA6YG, JA7BJA, JX1AK, JX5AZ, JY6UMS, OD5FB, VK30C, VK3BJB, W7MTL/P, ZL2AAW, ZL2BFU, ZM2KG, ZK2BD all on s.s.b.

Thirteen-year-old Chris Jones is alive and well and living in West Bromwich. He also bagged: EA3WZ, IT9FKS, JA3XKJ, W6CYU, all with a CX203 and 70 ft . end fed on twenty metres s.s.b.

We stop this 14 MHz revelry to bring you a news flash: a two metre $\log$ has been received. Guilty party is one David Noakes who resides at Cranbrook in Kent. Using an f.e.t. converter (i.f. $28-30 \mathrm{MHz}$ ), Trio JR310 and a 6-over-6 slot-fed he logged: F1AGY, F1CCP, F6CKU, G3AMF, G3FAB, G3RPZ, G3WKS/P (Rutland), G3YOA, G8BXX/P, G8CIW, G8CLY/P, G8FUF, G8FPR, G8GTD/P.

Derrick Dance (Roxburghshire), CR70A, PR40 pre-elector, a.t.u. and 264 ft . end fed sends in a nice log of stations heard on topband. On c.w.: DJ3CY, GI3YFY, GW3RTR, OHISJ, OLIAOH, OLIAQL, OL5AQK, plus several GM, OK1 and OK2 stations. On 160 metres s.s.b.: GM4BBL (Shetlands), GW3UCB, GW3ZQN, W4BIF, OE5KE.

At the other end of the scale, Derrick reports these on 28 MHz c , w.: DLOIGI (beacon station), F3NB, ZC4CY (new beacon), 3B8MS (beacon), 3D6AX, $4 \mathrm{X} 4 \mathrm{SK}, 9 \mathrm{HlCH}$. The s.s.b. signals on 28 MHz came from: CR6AM, CR6LF, CR7IZ, EA8GZ, EA8IB, LU8AJG, ZC4DS, ZE1AV, 4X4HF, 4X4NJ, 4Z4DZ, 4Z4LI, 9J2AY, 9J2TB and PY5UI on a.m.

Events during June include: June 9-10, four metre open contest; 10, d.f. qualifying round at High Wycombe; 16-17, Microwave Field Day; 17, WAB v.h.f. contest (phone); 23-24, First Topband contest; 24, d.f. qualifying round at Chelmsford. July 7-8, Jubilee v.h.f./u.h.f. contest.

## BROADCAST BANDS

Short Wave Reports by 15th of the month to Mslcolm Connah, 59 Windrush. Highworth, Swindon, Wiltshire, SN6 7DT.
Medium Waves Logs to Charles Molloy, 132 Segars Lane, Southport, PR8 3JG.
VHF/FM Reports to Simon David, c/o Practical Wireless, Fleetway House, Farringdon Street, London, EC4A 4AD.

## AMATEURBANDS

## Short Wave/VHF

Logs in alphabetical order please by 15th of the month to David Gibson, G3JDG, 12 Cross Way, Harpenden, Hertfordshire.

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## TAK DAVID ANDREWS

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## A series of simple transistor projects, using not more than twenty components.

THIS month's project describes a siren that utilises a voltage controlled relaxation oscillator. The output of the oscillator is fed to a very simple loudspeaker driver stage comprising Tr 3 and $\operatorname{Tr} 4$ connected as a super-alpha pair. Alternative output stages to provide higher power can be used provided they present a comparatively high input impedance.

## Circuit

Tr 1 and Tr 2 form the active part of the oscillator. They can be almost any type of general purpose pnp and npn transistor respectively; but make sure they are connected the right way round. The fre-


The complete circuit of the electronic siren.

## $\star$ components list



No. 50
ELECTRONIC SIREN


Suggested layout of components on a piece of veroboard.
quency of oscillation is set by the values of R 1 and Cl together with the potential at the base of Trl. In this instance the potential is determined by the network of R2, R3, R4 and R5 and is dependent on whether switch SWl is open or closed. C2 serves to delay the build up and fall of this potential and gives a smooth change of frequency that gives rise to the siren "wail".

Assuming that SWl is open and that the potential at Trl's base is low, Cl will charge up through R1 until the base/emitter junction' of $\operatorname{Tr} 1$ is forward biased. Not much charge is needed on Cl to arrive at this situation and it occurs very early on in the charging cycle. When Trl is turned on regenerative feedback occurs between Trl and Tr2 and the charge of Cl is shunted to negative through the circuit of Trl's emitter and collector and Tr2's base and emitter; at the same time the collector of Tr 2 is pulled down towards negative. Eventually base current into Trl ceases to flow and Cl is free to re-charge through R1 to repeat the cycle; the potential at Trl's base returns to its original state. If this potential was higher it would take more time for Cl to reach a charge whereby the emitter of $\operatorname{Trl}$ became more positive than the base and thus the frequency of oscillation is reduced. By closing SW1 we force the potential to increase but this does not happen immediately and the rise is slowed up by the time constant given by C2. The frequency of oscillation reduces slowly to a minimum value set by the potential divider effect of R4 with R5. When the switch is opened the potential reduces slowly as the charge on C2 is bled to negative by the momentary spikes of current drawn by Tr2 through R2. While this happens the frequency of oscillation slowly increases.


## Shoreham Transmitters

I was interested to read Mr. Woodward's letter in your March issue. He complains that our transmitters at Shoreham "cause countless harmonics all the way down the medium wave band". The frequencies used at Shoreham are 692 kHz (Radio 4), 1214 kHz (Radio 1), and 1484 kHz (Radio Brighton).

It is true that after the extensive wavelength changes last September a number of intermodulation products were identified but the situation was quickly rectified. G. H. Sturge ( $B B C$ Engineering Information Dept).

## Noises

. . . The fact remains that several miles from the transmitters, noises identifiable as Radio 1 can be heard loud and clear on 266 metres, 232 metres, and also on 1100 metres long wave on receivers ranging from small transistor radios to powerful communications receivers. The strength of these signals varies from day to day, and on some occasions I have heard Radio 1 very weak on 247 metres, and very strong on 323 metres.

In addition, Radio Brighton is always clearly audible on approximately 530 and 560 metres medium wave, as well as the intentional 202 metre wavelength.

My complaints have also been supported by the Horsham-based Southern Independent Radio Association, and the Brightonbased Southern. Free Radio Campaign, both of which are responsible organisations, who are trying to improve the standards of broadcasting in this country.

Regarding the possibilities of covering the whole country on medium wave with a few kilowatts, the BBC should try using horizontal long-wire aerials instead of vertical masts. Using this method, the Independent Broadcasting Authority's test broadcasts on 539 metres medium wave cover large parts of the country, with a power of one kilowatt, despite the fact that a powerful Dutch pirate station is on 538 metres, directly underneath.

The IBA's transmitter is also audible in parts of Holland, over the aforementioned Dutch station. 1 also know for a fact that many on-shore pirate radio stations, notably Radio Jackie of London and Focus 240 of Worthing cover distances of as much as 30 miles, with the 'massive" transmitter power of 7 watts! They also use efficient, long-wire aerials. $\mathbf{P}$. Woodward (Lancing, Sussex).

## Corrective Action

. . . Following Mr. Woodward's further letter, we have carried out some further checks and measurements in the Brighton area. We found two spurious signals spaced about 80 kHz above and below the frequency of Radio $1(1214 \mathrm{kHz})$. These signals were intermittent and of comparatively recent origin, and they were only present when a certain combination of enuipment was in use at Shoreham. The necessary corrective action has now been taken, and we are grateful to Mr. Woodward
for drawing our attention to these signals.

We did not find any trace of a transmission of Radio 1 on Long Waves, or of Radio Brighton on 530 or 560 metres. But it may be worth pointing out that with a receiver i.f. of around 465 kHz , the 'image' appears at approximately the frequencies mentioned, because

$$
\begin{aligned}
& 1484-(2 \times 465)=554 \mathrm{kHz} \\
& \text { (i.e. } 542 \text { metres) } \\
& 1214-(2 \times 465)=284 \mathrm{kHz} \\
& \text { (i.e. } 1058 \text { metres). }
\end{aligned}
$$

Such 'signals' are of course entirely the product of the superhet receiver, and do not exist as transmissions at all.

There is no point in going into great detail about the problems of medium-wave transmitting aerial design here, which are many and complex. I am afraid that the horizontal long-wire aerial which Mr. Woodward suggests is not suitable for serving very large areas by day and by night.

The IBA test transmissions are not made from such an aerial, and their relatively large coverage is due mainly to the high earth-conductivity of the area adjacent to the River Thames. Nevertheless the station does not serve large areas of the country nor is it intended to do so.

We understand that recordings made in Holland reveal the IBA signal as only just detectable in the presence of the pirate. It should also be remembered that there is a limit to the degree to which the service area of a station can be enlarged by increasing the power or using a different type of aerial, and the power necessary to serve a given area depends on very many factors, not the least of which is the level of interfering signals which must be overridden. G. H. Sturge (BBC Engineering Information Dept).

## TAKE 20-Electronic Siren

-continued from page 260

## Operation

On initial switch-on it is possible to get no immediate high frequency oscillation and this problem can be overcome by temporarily shorting out R3 and when the audible tone is heard the shorting link can be removed. Varying the potential divider effect of R4 and R5 will set the minimum lowest frequency. The highest frequency, before complete cut off, is best set by Cl. R1 could be adjusted to vary the range of frequency but this is undesirable
as it affects the working points of the two transistors.
The output is taken from the junction of R1 with Cl and this is a very high impedance point and if too much current is drawn by an external circuit Cl may never reach a charge at which regenerative triggering can occur. Hence the need for a reasonably high input impedance to the amplifier stage. The output waveform is a sawtooth and its amplitude decreases as the frequency increases which is quite effective in compensating for poor low frequency response of small loudspeakers.

The estimated cost to build this project is $95 p$, plus cost of the speaker, Veroboard, and VAT.

## ON RECENT DEVELOPMENTS

COMPUTERS are steadily becoming increasingly involved in our lives. Latest event to highlight this was the official opening of America's first computer-controlled petrol station which also boasts on-line credit checks. The motorist can pay cash or offer a credit card; the mini-computer will accept either. However, just like its human counterpart, man, the computer is happy to accept cash but wary of the credit card. When such a card is offered the computer will immediately check and verify it with the central computer files held in Atlanta. The petrol station is in the suburbs of Los Angeles. If the credit card offered to the computer is out of date, or has been stolen, then the computer will simply "pocket" the card and will refuse to return it. Now there's something you'd scarcely credit.

## ONE STEP AHEAD

A quick note to exponents of ESP (extra sensory perception). There is to be an International conference on "Electrical Signals from the Brain". It is to be held at Oxford University, August 28-30, 1973. Anyone got any brain waves?

## DEAF-BLIND HELP

In Hanover, there is a Centre for deaf-blind people. How do such unfortunate humans communicate? Feel or touch is an obvious answer. They can, for example, read Braille by feeling the raised dot impressions which make up the Braille alphabet. They cannot, of course, either see or hear television nor can they hear or answer a telephone call.

An ingenious system has been installed at the Hanover Centre which enables deaf-blind people to communicate directly with each other, over great distances-theoretically, anywhere in the world. Key to this system is the Braillomat. This machine produces a strip of paper with Braille alphabet letters or dot characters raised on it. Thus the deaf-blind person has only to run the strip of paper through his/her fingers to "read" the message. The Braillomat has a keyboard rather like a typewriter, but it is extremely simple in that it has only six keys-one key for each of the dots in the Braille alphabet pattern. The idea is that by connecting two of these machines together, two deaf-blind persons can "chat" happily. The tape output, incidentally, is rather like teletype or teleprinter tape.

It doesn't take too much imagination to see the potential here. Each deafblind person with a Braillomat could
contact other deaf-blind people by using the machines in conjunction with the telephone. But wait; how will they hear the telephone ring? In practice they don't have to because the inventors have thought of this problem. Their answer is a tiny unit which is carried on the personrather like a radio paging bleeper. Instead of emitting a warning bleep when the telephone rings, the unit simply vibrates. What a wonderful application of electronics this is.

## THE "R.O.M."

Big international exhibitions are heavy going but often there are one or two very interesting items which crop up. At this year's Mesucora exhibition in Paris, one company exhibited a digital multimeter which incorporated a digital thermometer. Thus, in addition to the usual multimeter capability of measuring voltage, current and resistance, it offered the measurement of temperature also.
Unfortunately, as temperature rises, the transducer or sensor gives an output which does not follow a linear law and therefore cannot "change" degrees $C$ directly into digital reading.

The solution used was to employ a ROM (read-only memory). This memory can be programmed to take certain correcting actions. For example; if the sensor was measuring $500^{\circ} \mathrm{C}$ and this, when fed directly to the digital display gave a reading of, say, 537, then the ROM could be programmed to compensate for this. Whenever a $500^{\circ} \mathrm{C}$ signal came in, the ROM would alter things by subtracting 37 from the answer before it was displayed.

Again, perhaps $600^{\circ} \mathrm{C}$ might read (fed direct) 671 and thus the ROM's programme would tell it, "If 600 in, subtract 71 to give 600 display". Thus the whole programme takes account of any input and processes it so that the "correct" temperature reading is shown digitally.

## INTEREST IN "ERNIE"

In the May article, the story about "ERNIE" might have implied that the name was derived from Ernie Harrison and possibly that Racal Electronics were connected with the Premium Bond selection equipment We would like to record that the electronic equipment was developed by Plessey and that ERNIE is derived from Electronic Random Number Indicating Equipment, the original version being developed at the Post

Office Research Station, Dollis Hill, London. Mr. Harrison is Chairman of the Southern Region Industrial Savings Committee; Racal are not directly connected with ERNIE.

## FLAT-SCREEN TV

Last month I reported thin readout devices and suggested flat screen television. I have just come across a flat-screen cathode ray tube (c.r.t.) or, to be more precise, two inches deep. Its function (at present) is to display alpha numeric characters/messages and it does not use either electron guns or deflection plates. It is digitally addressed and the display forms 512 characters in a $5 \times 7$ matrix of dots. The screen measures $5.4 \times 3.8$ inches. It has an area cathode which emits a sort of blanket of electrons as opposed to the narrow beam normally associated with the more conventional (and very much longer) c.r.t.
Beam forming is achieved by a plate with some 20,000 holes in it. This "forms" the electron blanket into individual beams. These beams then have to pass through four plates each similar to the beam forming one. These plates are the switching mechanism by which characters are formed. The first two plates (of the four) determine the position of the characters on the screen while the remaining two determine the column and row.

A 75 V positive potential will accelerate electrons through the holes in the switching plates, while a 50 V negative potential will cut off the beam completely.

After travelling through the switching plates the electron beam(s) now pass through a final plate used for focusing. They then strike the phosphor screen thus creating an illuminated image. The very small thickness of this unique c.r.t. ( 2 in .) has numerous obvious advantages.

Other advantages are also present which may not, at first sight, seem quite so obvious. Because the electron beam has such a very short distance to travel, the display is virtually immune to magnetic interference.
Note: This display, although it has been exhibited and is not an industrial secret, is not yet on the market although its appearance is believed to be imminent.
Cimbers

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PART FOUR

JUREK BUDEK*



## -continued from the June issue

As the load increases, the suppression of radio frequency interference becomes more difficult. Many sophisticated and very effective circuits have been published and are on the market, but very often they are prohibitive to the normal user because of their high cost.
The suppression filter used in this article is very simple, inexpensive and quite effective. It is suitable for total loads up to 900 watts, and consists of one bifilar wound ferrite core filter choke and $0 \cdot 1 \mu \mathrm{~F}$ capacitor as shown in Fig. 23.


Fig. 23: The suppression circuit used in the mains input suppiy to the lamp dimming controi unit.

The normal symmetrical currents generate magnetic fields which are self cancelling. However, symmetrical currents, caused by radio frequency interference are attenuated.

## Assembly

Pay attention to the correct polarity of capacitors, diodes, transistors and triacs. Mount the pulse transformer in correct way, i.e. to avoid sideway fixing.

The three core mains lead enters at the rear of the chassis (Fig. 20, page 54, May issue) and is connected to the junction terminal block.

Thereafter, the connections to the junction terminal block are as indicated in the printed circuit board wiring layout diagrams in this issue. It is pointed out that the mains 'earth' lead (E terminal) should preferably be taken directly to the control unit chassis. In Fig. 20 (May issue) this was shown soldered to the 'earth' side of the printed circuit board layout.

When the printed circuit board is mounted in the chassis metalwork, the 'earthy' side of the foil is connected to the metalwork via the mounting spacers, thus ensuring continuity of the chassis-p.c. foil earth system.
Therefore, the mains earth lead (common) and the earth lead from tag 1 on the lamps output socket SK1 may be conveniently taken to a chassis earth tag which may be mounted under one of the metal bolts that hold socket SK1 in position.

The socket SK1 is wired otherwise as shown in Fig. 20.

The three triacs, CSR3, CSR4 and CSR5, are in all instances bolted directly to the common heat sink


Fig. 24: Rear view of the P.W. Tricolour. The DIN piug and socket are the audio signal feed points-to amplifier output and loudspeaker.

[^6]
Fig. 26 : Component layout for the zero voltage version of the P.W. Tricolour on
the Mk II printed circuit board. Refer to the April' 73 issue for component detalls.


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Fig. 27: Cabinet details for the
P.W. Tricolour.

Fig. 28 : Details of the base unit for the lamps.



Fig. 29 : Front panel and chassis metalwork detalls for the lamp dimming version. The heat sink is identical to that given in the May issue. Note that the two holes marked ' $K$ ' on the front panel should have been given as ' $H$ '.
using 4BAx ${ }^{1}{ }_{4}$ in. bolts and nuts. No insulating washers are used. The heat sink is held in position on the printed circuit board by $4 \mathrm{BAx}^{1}{ }_{4} \mathrm{in}$. bolts and nuts.
The lamp dimming controls, VR2, VR3 and VR4 are wired as shown in Fig. 25, the left hand tag on each control is unused. These tags should be carefully positioned so that they do not touch or short to adjacent components.
The Bulgin socket (SK1) and plug (PL1) are numbered on their mouldings to conform with Fig. 20.

Having completed the units and checked the lamp wiring to the multiway lamps socket-1-Earth, 2Common, 4-Bass, 6-Middle and 8-Treble, check if any of the lamps come on, after the supply has been switched on. If so, there may be a component wrongly positioned in the filter circuit or a solder bridging around the triac or thyristor. If no lamp
lights, press on the override switches; the lamps should come on, and if not, change over the lamps, check interconnecting wiring and switching circuits i.e. triac and thyristors.

Finally, play some music. If any of the lamps do not work, check all the components in the relevant filter circuit for dry joints etc.

It should have been pointed out in the captions to Figs 24 and 27 (part 3, last month) that each division of the graticule scale represented 1 cm . In the list of components applicable to Fig. 25 the value of R1 (page 184) should have been given as $15 \mathrm{k} \Omega$.

It would appear that in some copies of the May issue, one of the I.C. connections (IC1) in Fig. 9 was missing. The missing number is ' 1 '. On the same diagram a figure ' 13 ' appears between the middle and treble outputs. This should be ignored.

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| $220 \mu \mathrm{~F} \quad 6 \frac{1}{2} \mathrm{p}$ | $68 \mu \mathrm{~F} \quad 6 \frac{1}{2} \mathrm{P}$ | $150 \mu \mathrm{~F} \quad 10 \mathrm{p}$ |
| $330 \mu \mathrm{~F} \quad 6 \frac{1}{2} \mathrm{P}$ | $150 \mu \mathrm{~F} \quad 8 \mathrm{P}$ | $220 \mu \mathrm{~F} \quad 11 \mathrm{p}$ |
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| $68 \mu \mathrm{~F}$ - $6 \frac{1}{2} \mathrm{P}$ |  |  |
| $150 \mu \mathrm{~F} \quad 6 \frac{1}{2} \mathrm{P}$ | 25 VOLT |  |
| 470, F - 11 p | $10 \mu \mathrm{~F} \quad 6 \frac{1}{1} \mathrm{p}$ |  |
| $680 \mu \mathrm{~F} \quad 13 \mathrm{p}$ | $22 \mu \mathrm{~F} \quad 6 \frac{1}{2} \mathrm{P}$ |  |
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|  | $220 \mu \mathrm{~F} \quad 10 \mathrm{p}$ | $1 \mu \mathrm{~F}$, $6 \frac{1}{3} \mathrm{P}$ |
|  | $470 \mu \mathrm{~F} \quad 13 \mathrm{p}$ | $2 \cdot 2 \mu \mathrm{~F} \quad 6 \frac{1}{2} \mathrm{P}$ |
|  | $680 \mu \mathrm{~F} \quad 20 \mathrm{p}$ | $4 \cdot 7 \mu \mathrm{~F} \quad 6 \frac{1}{2} \mathrm{P}$ |
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 GP280. GARRARD GCM21, and 22, GC823, GKg25 and 26, GCM21T and 22T, GCM24T GC823. GK825T and 26T, GCM31, GCE36, GCS3B, GC838, K840A, K841B ETC, ALL AT 68p (7p). SAPPHIRE 82p (7p), DODBLE DLAMOND 8TYLI: GRme dia BT/LP tip each side; no 78) All types si.85 (7p). SAPPHIRE DOUBLE ST/LP, TIPPED 82p (7p). DIAMOND STYLII GOLDRING REPLACGMENTS: 9800 , G800H \& G850, 22.25 each. G800E, 88.75 ( 8 p each). PICK-UP WIRE: super thin flex screened, sheathed, TWIN 7p. per yard, 4 -core 18p per yard (either up to $6 \mathrm{yds}, 7 \mathrm{p}$ ). Over, charges paid. Micropronis: CR YRTAL: LAPEL 18 ,' clip/hand, lead 3 hand 58 p (9p), 50 K 0 , Excen pluts 87.87 ( 8 p ) Cardiod Ball, $50 \mathrm{~K} / 600 \Omega$ built in volume control, on off 8 witch, special $20 f t$ lead, the best value anywhere at $86 \cdot 60$ (33p); Uni-dir mesh ball $50 \mathrm{~K} / 600 \Omega$ jack plug cable, adaptor, 25.28 (33p): Omni-dir. Ball mesh, 50 K , cable adsptor, jack plug, 44.29 (83p each), SPEAKERS; Very popuiar $12^{\prime \prime}$ ROUND, fitted tweeter, 3,8 or 16 ohms (state which) \&2.06 (33p) or pair for stereo st-65, charges paid. 8MALL $2 \jmath^{\prime \prime}, 3 \Omega, 8 \Omega$ or $64 \Omega$ (state which) $4 l^{2}$ (9p). More speakers in List. EEADPHONES: High resistance $2000 \Omega$ adjustable $61 \cdot 10$ (11p), EARPIECES: with lead and min. 2.5 mm or $3 \cdot 5 \mathrm{~mm}$ (stato Which) jack plug. MACNETIC 18p. CR YBTAL (3.5mm plug only) 88 p (up to 6 for $9 p$,
 highest quality, fully guaranteed si-29 (11p). TRAR 17 p (Up to 12 for 7p). CORNECTLAG WIRE: Packs of 5 colis, each coil Syds. Absts. cols. SOLID CORE 16p (7p). FLEXIBLE CORE $18 p$ ( $9 p$ ). RETRACTABLE FLEXIBLE LEADS (CURLIES) with phono plug each end, or phono plug one end, phono socket at the other (state which) $6 \mathrm{ft}, 30 \mathrm{p} ; 12 \mathrm{ft}, 53 \mathrm{p}$ (either 7p). VIBRATORS: 12v/4 pin non-synch 121HD4, $27^{\prime \prime}$ ex. pins, 83p. 8AME but $3 t^{\prime \prime}$ ex. pins, U8A, $16 \frac{1 p}{}$. $12 \mathrm{v} / 7$ pin synch. (12sR7) $2 \mathrm{t}^{\prime \prime}$ ex. pins, 72 p ( 9 p any type). MAINS BATTERY ELIMIEATORS: Two models, both O40v A.C. Input. Buit mort radios and cassette recorders/players with front switch, pllot lights, mains and output leads.
Model $S C, 3,6.71$ and $9 v ~$
400 mA . stahllised with multi-output adaptor 45.50 ( 35 p ). Model $\mathrm{SC}_{1}$, 3, 6 . 7 f and 9 v 400 mA , stahllised with muiti-output ad
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6000.08 6000.08
8000.11 8000.11
10000.12 1000
1200

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|  |  | Descriptlon | Pri |
| :---: | :---: | :---: | :---: |
| 01 | 120 | Glass Sub-Mia. General Purpose Gernianium Dlodes | $\begin{aligned} & 40 \\ & 0.55 \end{aligned}$ |
| 0 | 60 | Mixed Gerinanlum Transletors AF/RF | 0.65 |
| U |  | Germanium Gold Bonded Bub-MIn. like OAS, OA47 |  |
| U | 40 | Germantum Tranalstore like OC81, ACl28 |  |
| U 5 | 60 | 200ma $81 \mathrm{nb}-\mathrm{M} \mathrm{m}$. 8 ilicon Dlodes |  |
| U 6 | 30 | Sil. Planar Trans. NPN like BSY95A. 2N706 | 5 |
| U | 16 | 811. Reetiflers TOP-HAT 750 mA VLTG. RANGE up to 10 | 5 |
| 08 | 50 | 811. Planar Dioden DO-7 Glass 280 mA like OA200/202 | 0.55 |
| U9 | 20 | Mired Voltages, 1 Watt Zener Dlodes | 5 |
| U10 | 20 | BAY50 charge etorage Dlodes DO-7 Glase |  |
| U11 | 28 | PNP SII. Plense Trens. T0-6 like 2N11s2, 2N2904 |  |
| T12 | 12 | Eillcon Rectifiers Epory 600 mA up to 800 PIV | 55 |
| -13 | 30 | PNP-NPN Sil. Translators OC200 \& 28104 | 0.65 |
| U14 | 150 | Mixed sillcon and Germanium Dlodes | 5 |
| U15 | 25 | NPN 811. Planar Trana. TO-6 like BFY51, 2N897 | 5 |
| U18 | 10 | 3 Amp silcon Rectifiers 8tud Type up to 1000PIV |  |
| 017 | 30 | Germanlum PNP AF Transistors TO-5 like ACY 17-22 | 0.85 |
| U18 |  | 6 Amp Snlicon Rectifers BYZ13 Type up to 600 PIV | 86 |
| U19 | 25 | Bilicon NPN Transistors like BCl08 |  |
| రิ20 | 12 | 1.8 Amp Silicon Rectiflers Top Rat up to 1000 PIV |  |
| [21 | 30 | AF. Germanlum Alloy Transintors 20300 Series \& OC71 |  |
| U23 | 30 | MADT's like MHz Series PNP Transistors |  |
| U24 | 20 | Germsnium 1 Amp Rectifers GJM Series up to 300 PIV | $0 \cdot 56$ |
| U28 | 25 | 300 MHz NPN Silicon Transistore 2N708, B8Y27 |  |
| U26 | 30 | Fast Ewitching silicon Diodes like IN914 Micro-Min. |  |
| U27 | 12 | NPN Gormanlum AF Translators T0-1 like ACl27 |  |
| O29 | 10 | 1 Amo BCR's T0-5 can, up to 600 PIV CRB1/25-600 |  |
| U30 | 15 | Plastic Silicon Planar Trans. NPN 2N2826.. | 0.85 |
| U31 | 20 | silicon Planar Plastic NPN Trans. Low Nolse Amp 2N3707 |  |
| U32 | 25 | Zener Diodes $400 \mathrm{~mW} \overline{\mathrm{D}} 0-7$ cane $3-18$ volts mixed |  |
| U33 | 15 | Plastic Case 1 Amp Silicon Rectifera IN 4000 Serles |  |
| U34 | 30 | Snicon PNP Alloy Trans. TO-5 BCY 26 28302/4 |  |
| U35 | 25 | Silicon Planar Tranaistors PNP TO-18 2N2906 ... |  |
| U96 | 25 | Gilicon Planar NPN Translators TO-6 BFY $00 / 61 / 62$ |  |
| 037 | 30 | Sillcon Alloy Trandintors 80-2 PNP OC200, 28322. |  |
| U38 | 20 | Frat 8witching gllicon Tranm. NPN $400 \mathrm{MHz} 2 N 3011$ |  |
| U39 | 30 | RF. Germ. PNP Transistors 2N1903/5 TO-5 |  |
| $\overline{\text { U40 }}$ | 10 | Dual Transistors 6 lead TO-6 2N2060 |  |
| U41 | 25 | IFF Germanlum Transiztore T0-1, 0C46. NKT72 ... |  |
| O42 | 10 | VGF Germanium PNP Transiators T0-1 NKT667, AF117 |  |
| U43 | 25 | (il. Trans. Plantlc TO-18 A.F. BC113/114 |  |
| U44 | 20 | S!1. Trans. Plastic TO-5 BC115/116 |  |
| U45 |  | 3A SCR. T06600 to 800PIV | $1 \cdot 10$ |

## QUALITY TESTED SEmiCONDUCTORS

| Pak |  | Price Ep |
| :---: | :---: | :---: |
|  | 20 Red spot tranalstors P | 0.85 |
| Q2 | 16 White spot R.F. trangi |  |
| Q3 | 40 Cl 77 type trans |  |
| Q4 | 6 Matched translstor |  |
| O5 | 4 OC 75 transistors |  |
| Q8 | 5 OC 72 transistors |  |
| Q7 | AO 128 transistors |  |
| Q8 | AC 126 transist |  |
| Q9 | OC 81 tyde translators |  |
| Q | OC 71 type tranulator |  |
| Q11 | 2 AC 127/128 Compiementary PNP/NPN |  |
| Q12 | 3 AF 116 type transistors |  |
| Q13 | AF 117 type transistors |  |
| Q14 | OC 171 H.F. type tranastors |  |
| Q15 | 7 2N2926 Sil. Epoxy transistors mixed colours | 0 |
| Q16 | 2 GET880 low noise Germanium transistors |  |
| Q1 | 5 NPN $2 \times 8 T .141$ d $8 \times$ BT. 140 |  |
| Q18 | $\begin{aligned} & \mathrm{MADT} \text { M } 2 \times \operatorname{MAT} 100 \& 2 \times \mathrm{M} \\ & 120 \end{aligned}$ | 0.85 |
| Q19 | 8 MADT's $2 \times$ MAT 101 \& $1 \times$ MAT |  |
| Q20 | 4 OC 44 Germenlum transistora A.F. | 0.85 |
| Q21 | 4 AC 127 NPN Germantum tranalstora | . 0.85 |
| Q22 | 20 NKT transistorn A.F. R.F. coded | 0.55 |
| Q23 | 10 OA 202 Blicon diodes sub-min |  |
| Q24 | 8 OA 81 diodes |  |
| Q25 | 18 IN914 8ilicon dlodes 78 PIV | 0.55 |
| Q26 | 8 OA95 Germanium diodes IN 89 |  |
|  | 210 A PIV silicon rectifers 18425 | 0.85 |
| Q28 | 2 sllicon power rectiliers BYZ 13 | $0 \cdot 55$ |
| Q29 | $\begin{aligned} & 4 \text { sillicon translators } 2 \times 2 \text { N698, } \\ & 1 \times 2 \mathrm{~N} 697,1 \times 2 \mathrm{~N} 698 \times \ldots \ldots \end{aligned}$ | $0.56$ |
| Q30 | 7 Sillicon awitch transistora 2N706 NPN | $0.65$ |
| Q81 | - Bllicon mwitch translators 2N708 NPN | $0.55$ |
| Q32 | 3 PNP Bilicon traneintora $2 \times 2$ N1131, $1 \times 2$ N1132 | $\therefore 0.65$ |
| Q33 | 3 sillcon NPN transitors 2N17 |  |
| Q34 | 7 silticon NPN translatora 2 N2369, 500 MHz (code P397) | $0.55$ |
| Q35 | 3 Bullcon PNP TO-5. $2 \times 2$ N2904 \& $1 \times 2$ N2 | $0.55$ |
| Q3A | 7 2N3846 TO-18 plartic 300 MHz NPN | N 0.65 |
| Q37 | $82 N 3053$ NPN Bllicon trnnsintors | 3 |
| Q38 | 7 NPN transistors $4 \times 2 N 3703, ~ 子 \times$ $2 N 8702$ | $0.88$ |

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| 2N3819 | 81p | 2N5458 | 35p |
| :---: | :---: | :---: | :---: |
| 2N3820 | 85p | 2N6459 | 44p |
| 2 N 3821 | 39p | BFW10 | ${ }^{86}$ |
| 2 N 3823 | 81p | MPF105 | 11 |

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| :--- | :--- |
| UIC20 $=12 \times 7420$ | 0.65 |
| 0.55 |  |
| $1 C 30=12 \times 7430$ | 0.55 | IC30 $=12 \times 7430$ IC41 $=5 \times 7441$ $\begin{aligned} \mathrm{UIC41} & =5 \times 741 \\ \mathrm{UIC} & =5 \times 7442\end{aligned}$ $\mathrm{UIC43}-5 \times 7443$ $\begin{array}{lll}1044=5 \times 7444 & 0.55 \\ 0.55\end{array}$ U1C4 $=5 \times 7445 \quad 0.55$

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| SN7400 | $0 \cdot 17$ | 0.16 | 0.18 | SN7460 | 0.17 | 0.16 | $0 \cdot 18$ | 8N74141 $0 \cdot 74$ | 2.71 | $100+$ |
| SN7401 | $0 \cdot 17$ | $0 \cdot 18$ | 018 | SN7451 | $0 \cdot 17$ | 0.16 | 0.18 | SN74145 81 -65 | 21.54 | 81.48 |
| SN7402 | $0 \cdot 17$ | $0 \cdot 16$ | 0.18 | 8N 7453 | $0 \cdot 17$ | 0.16 | 0.18 | SN74150 23-80 | 28.97 | 22.75 |
| QN7403 | $0 \cdot 17$ | 0.18 | 0.18 | 8N7454 | $0 \cdot 17$ | $0 \cdot 16$ | 0.18 | SN74151 41 -10 | 81.05 | 0.99 |
| EN7404 | $0 \cdot 17$ | $0 \cdot 16$ | $0 \cdot 13$ | SN7460 | 0.17 | $0 \cdot 16$ | 0.18 | SN74153 ${ }^{\text {d }}$-82 | 21.21 | 21.05 |
| 8N7405 | $0 \cdot 17$ | $0 \cdot 16$ | 018 | SN7470 | 0-82 | $0 \cdot 29$ | 0.27 | 8N74154 £1. 68 | E1-87 | 81.76 |
| $8 N 7406$ | 0.39 | $0 \cdot 84$ | 0.31 | 8N7472 | $0 \cdot 32$ | $0 \cdot 29$ | $0 \cdot 27$ | SN74155 21.64 | 21-48 | 81.32 |
| SN7407 | 0.89 | 0.34 | $0 \cdot 31$ | 8N7473 | $0 \cdot 41$ | 0.39 | 0.35 | 8N7415621-54 | \$1.48 | \$1.82 |
| BN7408 | $0 \cdot 20$ | $0 \cdot 19$ | $0 \cdot 18$ | SN7474 | 0.41 | $0 \cdot 39$ | $0 \cdot 85$ | SN 74157 e2.08 | 21.98 | 21.87 |
| GN 7409 | 0.20 | 0.19 | $0 \cdot 18$ | QN7475 | $0 \cdot 50$ | 0.48 | 0.46 | 8N74160 81.98 | ¢1.87 | 81.76 |
| EN 7410 | $0 \cdot 17$ | $0 \cdot 16$ | 0.18 | SN7476 | 0.44 | 0.48 | 0.48 | SN7418181.98 | 81.87 | 81.76 |
| EN7411 | $0 \cdot 88$ | 0.87 | 0.88 | 8N7480 | $0 \cdot 74$ | $0 \cdot 71$ | 0 | SN74162 E4.40 | 24.18 | 48.85 |
| GN7412 | 0.89 | $0 \cdot 84$ | 0.81 | 8N7481 | 21-32 | 81.87 | 21-21 | GN74163 E4-40 | 14-13 | 48.85 |
| EN 7413 | 0.38 | $0 \cdot 28$ | 0.87 | SN7482 | 0.96 | 0.85 | 0. | SN74164 28.48 | 28.37 | 28.81 |
| SN7416 | 0.48 | 0.44 | 0.42 | SN7483 | 81.21 | 21.18 | 21.05 | SN74165 22.48 | 22.42 | 42.87 |
| SN7417 | 0.48 | 0.44 | 0.42 | 8N7484 | 81.10 | 81.05 | 0.98 | SN74166 23.85 | 48.58 | $28 \cdot 80$ |
| SN | $0 \cdot 17$ | 0.16 | 0.18 | SN 7485 | 28.96 | 88.85 | 23.74 | SN7417482.53 | 22.42 | 82.81 |
| SN 7422 | 0.65 | 0.58 | 0.50 | 8N7486 | 0. 35 | 0.34 | 0.33 | SN7417881-76 | 81.65 | 81-64 |
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| SN7425 | 0.65 | 0.68 | 0.50 | 8N7490 | 0.74 | $0 \cdot 71$ | 0.64 | SN74177 82-75 | 22.64 | 22.53 |
| SN 7428 | 0.50 | 0.48 | 0.44 | 8N7491 | \&1.10 | 21.05 | 0.99 | GN74180 £2. 20 | 81.78 | 21.54 |
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| 8N7437 | 0.71 | 0.88 | $0 \cdot 68$ | SN74100 | 211-89 | 21.76 | $81 \cdot 71$ | 8N74192 $22 \cdot 16$ | 22-09 | 22-04 |
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## Sinclair Project 60

## Now-the Z.50 Mk. 2

## with built-in automatic transient overload protection

When originally introduced. the Sinclair Z.50 proved how it was possible to design and produce a popularly priced modular power amplifier having characteristics to challenge the world's costliest amplifiers. Many thus ands of $Z .50$ 's are now giving excellent service day in, day out. But we have also learned that constructors do not always use their $Z .50$ 's ideally. That is why we have introduced modifications whereby risk of damage through mis-use is greatly reduced and performance further enhanced. The Z.50 Mk. 2 has improved thermal stability, more accurately regulated D.C. limiting to ensure more symetrical output voltage swing and clipping and still less distortion at lower power. Z.50 Mk. 2 is compatible with all other Project 60 modules. and may be incorporated to advantage in existing systems. Eleven silicon epitaxial planar transistors are now used. two more than in the original $Z .50$; circuitry has been re-designed, making this versatile high performance amplifier better than ever.
Z. 30 the power
amplifier for quality and economy

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free manual
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The $\mathbf{Z . 3 0}$ provides excellent facilitios for the constructor requiring a high fidelity audio system of less power than that available from 2.50 s. Using a power supply of 35 volts, $Z .30$ will deliver 15 watts RMS into 8 ohms, or 20 watts RMS into 3 ohms using 30 volis. Total harmonic distortion is a fantastically low $0.02 \%$ at 15 watts into 8 ohms with signal to noise ratio better than 70 dB unweighted. Input sensitivity 250 mV into 100 K ohms. Size $80 \times 57 \times 13 \mathrm{~mm}\left(3 \frac{1}{6} \times 2 \frac{1}{4} \times \frac{1}{2}\right) 230.2 .50$ and 2.50 MK. 2 modules are compatible and interchangeable


## Brilliant new technical specifications

with free
manual
£5.48
Input impedance $100 \mathrm{~K} \Omega$
Input (for 30 w into $8 \Omega$ ) 400 mV
Signal to noise ratio, referred to full $\mathrm{o} / \mathrm{p}$ at 30 v HT 80 dB or better
Distortion $0.02 \%$ up to 20 W at $8 \Omega$. See curve Frequency response 10 Hz to more than $200 \mathrm{KHz} \pm 1 \mathrm{~dB}$
Max. supply voltage 45 v ( $4 \Omega$ to $8 \Omega$ speakers)
( $50 \vee 15 \Omega$ speakers only)
Min. supply voltage 9 v
Load impedance-minimum: $4 \Omega$ at 45 v HT Load impedance-maximum : safe on open circuit

## Typical Project 60 applications

| System | The Units to use | together with | Units cost |
| :---: | :---: | :---: | :---: |
| Simple battery record player | 2.30 | Crystal P.U., 12 V battery volume control, etc. | ¢4.48 |
| Mains powered record player | 2.30, PZ.5 | Crystal or ceramic P.U. volume control, etc. | £9.45 |
| 12W. RMS continuous sine wave stereo amp for average needs | $\begin{aligned} & 2 \times Z .30 \mathrm{~s} \text {, Stereo } \\ & 60 ; P Z .5 \end{aligned}$ | Crystal. ceramic or mag P.U., F.M. Tuner, etc. | £23.90 |
| $25 W$. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers | $\begin{aligned} & 2 \times \mathrm{Z} .30 \mathrm{~s}, \text { Stereo } \\ & 60 ; \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U.. F.M. Tuner, Tape Deck. etc. | £26.90 |
| 80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. ( 60 W . RMS into 80 hms ) | $2 \times Z .50$ s, Stereo 60; PZ.8, mains transformer | As above | £34.88 |
| Indoor P.A. | Z.50, PZ.8, mains transformer | Mic., guitar, speakers. etc., controls | £19.43 |

F.M. Stereo Tuner (£25) \& A.F.U. ( $£ 5.98$ ) may be added as required.

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## the world's most advanced high fidelity modules

## Stereo 60 Pre-amp/control unit



Designed specifically for use on Project 60 systems, the Stereo 60 is equally suitable for use with any high quality power amplifier. Since silicon epitaxial planar transistors are used throughout. a really high signal-to-noise ratio and excellent tracking between channels is achieved. Input selection is by means of press buttons, with accurate equalisation on all input channels. The Stereo 60 is particularly easy to mount.
SPECIFICATIONS-Input sensitivities: Radio - up to 3 mV . Mag. p.U. 3 mV : correct to R.I.A.A. curve $+1 \mathrm{~dB}: 20$ to $25,000 \mathrm{~Hz}$. Ceramic p. U. - up to 3 mV : Aux - up to 3 mV . Outpur: 250 mV . Signal to noise ratio: better than 70 dB . Channel matching: within 1 dB . Tone controls: TREBLE +12 to -12 dB at $10 \mathrm{KHz} ;$ BASS +12 to- 12 dB at 100 Hz . Front panel: brushed aluminium with black knobs and controls. Size: $66 \times 40 \times 207 \mathrm{~mm}$

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## Project 60 Stereo F.M. Tuner



The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now. Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning. printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically. a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.
SPECIFICATIONS--Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz . Sensitivity: $7 \mu \mathrm{~V}$ for lock-in over fulf deviation. Squelch level: Typically $20 \mu \mathrm{~V}$. Signal to noise ratio: $>65 \mathrm{~dB}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}$ ( $\pm 1 \mathrm{~dB}$ ). Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation. Stereo decoder operating level: $2 \mu \mathrm{~V}$. Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S. maximum Operating voltage: 25-30VDC. Indicators: Stereo on; tuning. Size: $93 \times 40 \times 207 \mathrm{~mm}$.

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high fidelity amplifier


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sistor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC. 12 is compatible with Project 60 modules which would be used with the $Z .50$ and $Z .30$ amplifiers. Complete with free manual and printed circuit board.

## SPECIFICATIONS

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| 1T4 | 0.80 | 6BN6 | 0.60 | ${ }^{6} 36$ | 0.80 | 12AQS |
| 104 | 0.40 | $6 \mathrm{BQ5}$ | 0.88 | 6.37 | 0.45 | 12ATG |
| 106 | 0.75 | 6BR8 | 0.75 | $6 \mathrm{~K} 4 \Pi$ | 0.60 | 124T7 |
| 1v2 | 0.56 | 6B87 | 1.85 | 6K6(4T | 0.75 | 12AU8 |
| $1 \times 28$ | 0.58 | 6BW6 | 0.90 | 6 K 7 | 0.48 | 12AU7 |
| 2 D 21 | 0.40 | 6 BW 7 | 0.90 | 6K89 | 0.45 | 12AV6 |
| 8 A 4 | 0.45 | 6BX6 | 0.25 | 6 K 25 | 0.75 | 12AV7 |
| 8 Af | 0.75 | 6BZ8 | 0.45 | 6L6GT | 0.85 | 12AX7 |
| 884 | 0.40 | $6 \mathrm{C4}$ | 0.95 | $6 \mathrm{L7}$ | 0.45 | 12AY7 |
| 8 V 4 | 0.70 | 6CsGT | 0.85 | $6 \mathrm{L18}$ | 0.50 | 12B4A |
| Sbigy | 0.75 | 6CB6 | 0.40 | 6LD20 | 0.50 | 12BA6 |
| 5 U4G | 0.80 | 6CD6GA |  | 6N7GT | 0.55 | 12BA7 |
| 5 V 49 | 0.50 |  | 1.80 | ${ }^{6828}$ | 0.85 | $12 \mathrm{BE}{ }^{\text {d }}$ |
| SYJGT | 0.45 | 6CA7 | 0.60 | ${ }^{687}$ | 0.60 | 12BH7 |
| 5zs | 0.75 | 6CH6 | 0.60 | 6847 | 0.45 | 12 BY 7 |
| SZ4G | 0.45 | 6CL6 | 0.60 | 6897 | 0.45 | 12E1 |
| 8/80L2 | 0.00 | 60U6 | $0 \cdot 80$ | 68 Kk 7 | 0.50 | 12 K 5 |
| 6AB4 | 0.45 | 60w4 | 0.70 | $68 L 76 T$ | 0.4 | 12 K 7 O |
| 6AP4A | 0.60 | 6CY5 | 0.50 | 68N7GT | T0.45 | 12Q7G |
| 6AGS | 0.25 | 6 CY 7 | 0.75 | 6807 | 0.50 | 12887 |
| 64.77 | 0.45 | 6D3 | 0.55 | $68 \mathrm{R7}$ | 0.50 | 20D1 |
| 6AH6 | 0.60 | $6 \mathrm{DC6}$ | 0.80 | ${ }^{678}$ | 0.8 | 20 Pl |
| 6AJ8 | 0.80 | 6DK6 | 0.60 | 60477 | 0.70 | ${ }^{20 P 1}$ |
| 6AK6 | 0.40 | 6DQ8B | 0.75 | 6U69 | 0.75 | 20 Pb |
| BAK6 | 0.00 | 6DS4 | 1.25 | ${ }_{6}^{686}$ | 0.48 | 20 CO |
| 6AL3 | 0.48 | 6EA8 6EH7 | 0.68 | ${ }^{6 \times 6} 4$ | 0.40 | $25 L 66$ |
| 8ALS | 0.28 | 6EH7 | 0.80 | 6X5GT | 0.45 | 25Z4G |
| 6AM6 | 0.87 | 8EJ7 6EW6 | 0.70 | 6x8 ${ }^{\text {8 }}$ | 0.85 | $25 \mathrm{Z6G}$ |
| 6AgS | 0.48 | 6EW1 | 0.70 | ${ }_{6}{ }^{\text {Y }}$ BG | 0.80 | D0as |
| 6106 | 0.70 | ${ }_{6}^{6 F 1}$ | 0.76 0.75 | 7 74 | 0.75 | soa Es |
| BAR5 | 0.65 | ${ }_{\text {BFGG }}$ | 0.76 | 9BW6 | 0.70 | 80 Cl |
| 8 8A85 | 0.60 | 6F11 | 0.50 | 1002 | 0.60 | 80015 |



##  <br> $\qquad$

## Frat Quallty Fully Guarantood <br> 



| EF98 | 0.40 | G734 0.00 | PF818 | 1.00 | U801 | 0.85 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EF97 | 0.65 | HABC800.55 | PFL200 0 | 0.65 | U403 | 0.70 |
| EF98 | 0.75 | HK90 0.50 | PL38 | 0.40 | U404 | 0.70 |
| EF183 | 0.80 | KT66 8. 6 | PL36 | 0.65 | 0801 | 0.80 |
| EF184 | 0.85 | KT77 1.00 | PL81 | 0.50 | UABC80 |  |
| EF804 | 1.85 | KT88 8. ${ }^{\text {¢ }}$ | PLA9 | 0.45 |  | 0.40 |
| EK90 | 0.88 | N78 1.60 | PL83 | 0.45 | UAF61 | 0.70 |
| EL34 | 0.50 | PABC800.40 | PLS4 | 0.40 | UAF42 | 0.60 |
| EL36 | 0.50 | PC86 0.60 | PL302 | 0.95 | UB41 | 0.65 |
| EL41 | 0.75 | PC88 0.60 | PL504 | 0.75 | UBC41 | 0.65 |
| EL81 | 0.55 | PC92 0.50 | PL508 | 0.90 | UBC81 | 0.45 |
| EL83 | 0.50 | PC97 0.60 | PL509 | 1.10 | UBF80 | 0.40 |
| EL84 | 0.85 | PC900 0.48 | PL801 | 1.00 | UBFP9 | 0.40 |
| FL86 | 0.48 | PCC84 0.40 | PL802 | 0.08 | UBLI | 0.70 |
| ELS 6 | 0.40 | P(C85 0.40 | PM84 | 0.60 | UBL21 | 0.70 |
| EL90 | 0.48 | PCC88 0.65 | PY81 | 0.85 | UC92 | 0.45 |
| EL360 | 1.15 | PCC89 0.65 | PY93 | 0.88 | UCC8S | 0.45 |
| EL822 | 1.50 | PCC1890.60 | PY80 | 0.40 | UCF80 | 0.70 |
| ELLBO | 0.85 | PCC805 0.95 | PY81 | 8.80 | UCH21 | 0.60 |
| EM71 | 0.80 | PCC808 0.95 | PY82 | 0.85 | UCHen | 0.70 |
| EM80 | 0.45 | PCF80 0.00 | PY85 | 0.88 | UCH81 | 0.40 |
| EM88 | 0.60 | PCP82 0.25 | PY88 | 0.40 | UCL81 | 0.60 |
| EM84 | 0.15 | PCF84 0.60 | PY800 | 1.09 | UCL 82 | 0.25 |
| EM85 | 1.00 | PCF86 0.60 | PY800 | 0.47 | UCL8s | 0.65 |
| EM87 | 0.70 | PCF87 1.10 | PY801 | 0.60 | UF9 | 0.65 |
| ENP1 | 0.40 | PCP801 0.50 | PZ30 | 0.88 | UP11 | 0.60 |
| EY51 | 0.40 | PCF802 0.60 | gQvo |  | UF41 | 0.65 |
| EY80 | 0.75 | PCF805 0.90 |  | 2.85 | UF42 | 0.65 |
| EY81 | 0.40 | PCF806 0.75 | QQvo3. | -10 | UF48 | 0.85 |
| EY83 | 0.86 | PCF608 0.90 |  | 1.85 | UP80 | 0.25 |
| EY88 | 0.40 | PCH2000.70 | TT21 | 3.80 | UFB8 | 0.40 |
| EY87 | 0.48 | PCLA1 0.60 | TT2? | 4.00 | UF89 | 0.40 |
| EY88 | 0.48 | PCL82 0.85 | U18/20 | 0.75 | ULA1 | $0 \cdot 65$ |
| EZ40 | 0.50 | PCL83 0.65 | U25 | 0.85 | U186 | 0.48 |
| 昰41 | 0.75 | PCL84 0.45 | U26 | 0.85 | UN84 | 0.80 |
| Ez80 | 0.88 | PCL8 0.45 | U87 | 6.50 | UY12 | 1.00 |
| E281 | 0.20 | PCLA8 1.85 | U52 | 0.40 | UY41 | 0.4 |
| GT1C | 8.00 | PCL200 0.75 | U76 | 0.40 | UY82 | 0.89 |
| GY501 | 0.70 | PCL800 1.10 | U78 | 0.40 | U Y85 | 0.10 |
| GZ30 | 0.45 | PCL801 0.95 | U191 | 0.75 | W729 | 0.78 0.60 |
| GZ\$1 | 0.40 | PCL80s 0.60 | U201 | 0.50 | Y 63 | 0.75 |
| GZ32 | 0.60 | PD500 1.30 | U281 | 0.65 | $\underline{2759}$ | 3.60 |
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[^4]:    Fig. 7: Most of the components for the oscillator are mounted on the circult board shown below.

[^5]:    A vlew of the partially assembled casework showing the circult board mounted along the back whth input and output jacks and onloff switch at the lefthand side. The two batteries are located by the partition at the rlght. Pedal movement is restricted by the pedal housing, if constructed as In Fig. 7.

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