

RSC HIGH－FIDELITY STEREO PACKAGE OFFERS

## Firred units



Terms：Deposit
（Total $\mathbf{2 l 0 9}$ ．21）．
$\star$ TA12 AMPLIFIER $6 \cdot \overline{0}+6.5$ watt in veneered housing
＊Garrard spes mk ili Player unit on Plinth
$\star$ sonotone 9Tanc Ceramic Cartridge（diamond stylua） $\star$ Pair of dorchester Loudspeaker Units Or Deposit 58.40 and 9 methly．

Special payments of $\mathbf{8 6} \mathbf{3 2}$（Total $\mathbf{8 5 5 \cdot 2 8}$ ）．

Price
£59
Trans．Plastic Cover $£ 3.15$ extra．
Carr． 21.25.

PACKAGE AS ABOVE but with Garrard 3000 Antochanger and | Sonotone 日TA Ceramic Cartridge in |
| :--- |
| lieu of SP25．Carr． $21-25$. | Or Deposit $£ 7.50^{\prime}$ and 9 monthly payments 85.64 （Total $\mathbf{2 5 8 - 2 6}$ ）． Trans．Plastic Cover $\mathbf{2 3} 15$ extra．

Send S．A．E．for coloured brochure showing other money－
gaving offers．
＇YORK＇HIGH－FIDELITY 3 SPEAKER SYSTEM
 Impedance 15 ohms

COMPLETE KIT Carr．65p ＊Performance comparable with units costing congiderably more． Consists of（1）12in． 15 watt Bass unit with cast chassis，Roll rubber cone surround for ultra low resonance，and ceramic magnet．（2） 3 －way quarter
section series cross－over system（3） $8 \times$ 5in．high flux middle range speaker． （4）High pfficiency tweeter．（5）Appropriate quantity acoustic damping material．（6）Handsome Teak veneered cabinet．（7）Circuit and full instruc－ tions．Terms：Dep．$£ 4 \cdot 60$ and 9 monthly payments $\mathbf{\$ 2} \mathbf{4 7}$（Total 826 －88）． DEMONSTRATIONS AT ALL BRANCHES

## мя

Individual Ganged Controls：Bass，Treble，Volume and Balance，Printed circuit construction employing 10 Transistors plus Diodes．Output rating I．H．F．M．Frequency range $20-20,000 \mathrm{c} . \mathrm{p} . \mathrm{s}$ ．Bass Control $\pm 12 \mathrm{db}$ ． Treble Control $\pm 13 \mathrm{db}$ ．Selector switch for P．U．or Tape／Radio．For loudspeaker out－ put impedances of 3 to 15 ohms．For standard 200－250v．A．C．mains operation． Attractive Black and Silver finished metal facia plate and matching control knobs． COMPLETE EIT OF PARTS INCLUDING
FULLY WIRED PRINTED CIRCUIT and comprehensive wiring $\quad \leq| | .50 \quad \mathrm{Carr}$ ．
Or PACTORY BUILT IN TEAK FENEERED CABINET
$t 1.50 \quad \mathrm{Car}$
dep．£2．20 and 9 monthly payments $£ 1 \cdot 70$（Total $£ 17 \cdot 50$ ）．

## AUDIOTRINE HI－FI SPEAKER SYSTEMS

Consisting of matched 12 in ． 11,000 line 15 Watt 15 ohm high quality speaker，cross－over unit and tweeter．Smooth response and extended frequency range OR SENIOR 15 WATT INCL UDING OR SENIOR 15 WATT INCLUDING $£ 5.95$ $£ 6.95$ Carr．
 HF126 15，000 LTNE SPEAKER

## AUDIOTRINE HIGH FIDELITY SPEAKERS

Heavy construction．Latest high efficiency ceramic magnets Plasticised Cone surround D indicates Tweeter Cone providing extended frequency range up to 15,000
8.15 ohms．PLEASE STATE CHOICE．

|  | Exceptional performance at low cost． |  |  |  |  |  |  |
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| HFB08T | 8 | 10W | 2288 | HF120D | 12 | 15W | 24.98 |
| HF102D | $10^{\circ}$ | 10W | 23－40 | HF126 | 12 | 15W | £5．75 |
| HF120 | $12^{*}$ | 15w | £450 | HF126D | $12^{*}$ | 15W | £625 |

FANE 807T HIGH FIDELITY SPEAKER
A full range 8 in ． 10 watt unit for excellent sound quality，in suitable enclosure．Cast chassis Roll P．V．C．cone surround and long throw voice cone in fitted to extend high note response．Frequency range 25 Hz to 15 KHz ．Gauss 10,000 ．Impedance 3 or 8 －15 $\Omega$ ．

## HIGH FIDELITY LOUDSPEAKER UNITS

Cabinets latest atyle Satin Tesk veneer．Acoustically lined or filled
acoustic damping．Ported where sppropriate．Credit terms available
DORCHESTER（Illustrated）Size $16 \times 11 \times 9$ in．appr．Range $45-15,000$
c．p．s．Rating 8 －10 watts．Fitted High flux $13 \times 8 \mathrm{in}$ ． $\mathbf{1 9 . 4 5}$
Dual Cone speaker．Imp． 3 or 15 ohms．Carr． 40 p ． MONARCH Size $19 \times 10 \times 9 \mathrm{in}$ ．spprox．Rating 10 watts．Inc． $13 \times 8 \mathrm{in}$ ，speaker with highly flexible P．V．C．cone surround，long throw voice coil and 10,000 line magnet
High flux pressure tweeter．Handsome cabinet．Range $35 \cdot 20,000$ c．p．s．Imp． 8 ohm Gives smooth realistic sound output．See＇package offers＇tor
Gillustration
int illuatration
R．S．C．TAI2 MKIII $6.5+6.5$ WATT STEREO AMPLIFIER FULLY TRANSISTORISED，SOLID STATE CONSTRUCTION HIGE FIDELITY OUTPUT OF 6．5 WATTS PER CHANN Deaigned for optimum performance with any crystal
or ceramic Gram．P．U．cartridge，Radio tuner，Tape recorder etc．$t 3$ separate switched input sockets －Slide switeh $\star$ Separate Bass and Treble controls $3-15$ ohms $t$ For $200-250 \mathrm{v}$ ．A．C．Mains $\rightarrow$ Frequency
 Response $20-20,000$ c．p．s．－ 2 dB B Harmonic Distor－ （3） （3） 100 mV ．Output rating I．H．F．M．克 Handsome finish Facia plate \＆Knob
 Deposit $£ 3$ and 9 mthly pymts $£ 2.15$（Total 222.35 ）．Or in Teak vencer housing $\leq 23$ Deposit $£ 3$ and 9 mthly pymts $£ 2 \cdot 15$（Total $822 \cdot 35$ ）．Or in Teak veneer hous
Dep．$£ 3$ \＆ 9 mthly payments 82.55 （Total $£ 25 \cdot 95$ ）．Send S．A．E．for leaflet．

HI－FI SPEAKER ENCLOSURES MODERN DESIGN Teakk veneer Aniah．Acoustically lined．Sizes approx．Casr． 35 p ．per en．
JE8 Size $16 \times 11 \times 9$ in． $\mathrm{SE8}$ For optimum perform－
 8in．Hi－Fi speaker．SE12 For excellent
SE10 For outstanding results performance with 12 ent
Hi －Fi speaker and tweete with 10 in ． Hi －Fi
 $\stackrel{\text { Size }}{25} \times$

## R．S．C．BATTERY／MAINS CONVERSION UNITS

 TYPE BM1．An all－dry battery eliminator．Size $5 \ddagger \times 4 \frac{1}{2} \times 2 \mathrm{in}$ ． appron．Completely replaces batteries aupplying 1.5 v and 90 v ， COMPLETE KIT $\mathbf{6 3 . 2 5}$ ASSEMBLED READY $\mathbf{6 3 . 7 5}$WITH DIAGRAM
R．S．C．TA6 6 Watt HI－FI AMPLIFIER
200－250v．AC mains operated．Frequency Response $30-20,000$ c．p．s． －2dB．Harmonic Distortion 0．3\％at 1，000 c．p．s．separate Bass and $\mathrm{O}_{\mathrm{s}} \mathrm{S}$ O or Tape．Input selector switch．Output for $3-15$ ohm spkrs．Max．senstivity 5 mV Output rating I．H．F．M．Fully enclosed enamelled case， $9 \frac{1}{4} \times 2 i \times 5 \frac{1}{2}$ in．Attractive brushed silver finish facia plate
Complete kit of parts with full wiring diagrams and instructions．
40p． OR FACTORY BUILT WITH 12 MONTHS＇GUARANTEE $£ 10.95$


GUARANTEED．Impregnate
ies $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$ ．Screened
CHARGER TRNGORMERS $0-9-15 \mathrm{v}$ ）
 AUTO（Ste\％UPYSEep DOWN）TRANSFORMERS $0-110 / 120 \mathrm{v}, 200-230-250 \mathrm{v}$ ． $50-80$ watts $£ 1-10$ ，
150 watts，$£ 1.90250$ watts $£ 2.75 ; 500$ watts $£ 5.75$ OUTPUT TPANSPORMERS
Standard Pentode $5.000 \Omega$ or $7,000 \Omega$ to $3 \Omega \quad 50 \mathrm{p}$ $\begin{array}{ll}\text { Standard Pentone } 5.000 \Omega \text { or } 7,000 \Omega \text { to } 3 \Omega & 50 \mathrm{p} \\ \text { Push－Pull } 8 \text { watts EL84 to } 3 \Omega \text { or } 15 \Omega . . & 83 \mathrm{p}\end{array}$ Push－Pull 10 watts 6V6，ECL86 to $3,5,8$ or
$15 \Omega$ $\qquad$
Push－Pull EL84 to 3 or $15 \Omega 10-12$ watts．． $81-35$
Push－Pull Ultra Linear for Mullard 510 ，etc． Push－Pull 15－18 watts，sectionally wound $6 \mathrm{~L} 6, \mathrm{KT66}$ ，etc．，for 3 or $15 \Omega$
Push－Pull 20 watt high quality sectionally
wound EL34， 6 L6，K T66 etc．to 3 or $15 \Omega$

SELENIUM
RECTIFIERS RECTIFIERS
F．W．（Bridged）


## MIDGET CLAMPED TYPE $2 \$ \times 28 \times 2 \neq \mathrm{in}$ ．

 $250 \mathrm{v} .96 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .2 \mathrm{a} .$.$250-0.250 \mathrm{v} .60 \mathrm{~mA} 6.3 \mathrm{~V}$

FULLY SHROUDED UPRIGHT MOUNTING $250-0-250 \mathrm{v} .60 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .2 \mathrm{a}$ ．，0－5－6．3v．2a．$£ 1 \cdot 40$ $250-0 \cdot 250 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a},{ }^{0} 5-5-6 \cdot 3 \mathrm{v} .3 \mathrm{a}$ ． $22 \cdot 20$
$300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a}$, ， $0-5-6 \cdot 3 \mathrm{v} .3 \mathrm{a}$ ． $22 \cdot 20$ $300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a} ., 0-5-6 \cdot 3 \mathrm{v} .3 \mathrm{a}$ ．
$300-0-300 \mathrm{v} .130 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a}$.
c． $5 ., 6 \cdot 3 \mathrm{v}$ ． 1 a ． For Mullard 510 Amplifer $350 \cdot 0-350 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a}, 05-6-3 \mathrm{v} .3 \mathrm{a}$ ．
$350-0-350 \mathrm{v} .150 \mathrm{~mA}, 6 \cdot 3 \mathrm{v}, 4 \mathrm{a}, 0-5-6 \cdot 3 \mathrm{v} .3 \mathrm{a}$. $425-0-425 \mathrm{v} .200 \mathrm{~mA}, 6-3 \mathrm{v} .4 \mathrm{a}$ ．，c．t．， 5 v ． 3 a ． $82 \cdot 65$ $425-0-425 \mathrm{v}, 200 \mathrm{~mA}, 6$－3v． $4 \mathrm{a} ., 63 \mathrm{v} .3 \mathrm{a} ., 5$

R．S．C．Mk III SUPER 30 HIBH FIDELITY STEREO AMPLIFIER
A completely new design further improved in
both appearance and pertormance．Representing both appearance and pertormance．Representing
value far higher than the prices suggest．
Only high grade components by
leading manufacturers．
COMPLETE KIT OF PARTS Or FACTORY BUILT with 12 months guarantee．Dep．$£ 5.75$ and 9 monthly Or FACTORY BUILT in cabinet as illustrated．Dep． $\mathbb{\text { I }}$ and abinet as 9 monthly payments $£ 3.99$（Total $£ 42 \cdot 91$ ） LU0 J $\begin{aligned} & \text { gistors，Four Diod } \\ & \text { Four Rectifiers }\end{aligned}$ CONTROLS：PUSH－BUTTON SELECT（A） （4）Mono $L$（5）MonELECTOR（1）Disc（2）Radio（3）Tape （4）Mono L（5）Mono R（6）SPEAKER DIS．（7）Mains on／off．
－SATIN SILVER METAL FACIA with black lette ing．Black edged knobs with brg sulver centres． （ PUSH－BUTTON SELECTOR SWITCHING太 NEON INDICATOR
t JACK SOCKET FOR HEADPHONES IN SATIN CABINETED MODEL VENEERED IN SATIN TEAK．SUITABLE FOR ANY MODERN PICK－ UP CARTRIDGE CERAMIC Or MAGNEFND USE WITH THE BEST ANCILLA
MENT THAT CAN BEAFFORDED．
OUTPUT： 15 watts R．M．S．（Continuous）into 8 ohms． HUM R 10 watts R．M．S．（Continuous）into 15 ohms HARM NOISE－75dB Min．Vol．－65dB Full Vol． HARMONICDISTORTION $0 \cdot 1 \%$ at 1000 Hz 10 Watts
to 70 kHz BASS CONTROL $:+17 \mathrm{~dB}$ to -16 dB at 40 H FREQ．RESPONSE：-3 dB 7 Hz to 70 kHz BASS CONTROL：+17 dB to -16 dB at 40 H TREBLE CONTROL：+16 dB to -12 dB at 14 kHz CROSS TALK -58 dB SENSITIVITES：DISC MAS FOR 3 PAIRS OF INPUTS（1）PU（2）Radio．（3）Tape Amp Plus pair ABLE．Visually matches Super 30 Mklll ONLY £44．95

(b) PAIR OF HI-FI HEAD
(c) MATCHING DYNAMIC 'MIKE' (attached
(d) PAIR 50 WATT SPEAKERS Black Rexine covered Cabinets Size approx $18^{\prime \prime} \times 18^{\prime \prime} \times 8^{\prime \prime}$
(a)(b)(c) (d)
 Terms on Amps. Speakers and Head phone/Mike. Deposit $£ 15$ and 18 monthiy payments of $£ 4.00$ (Total
FAME ULTRA HIGH POWER LOUSPPFAKERSS

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FANE SPEAKERS 'POP' $25 / 212 \mathrm{in} .25$ WATT Dual Cone $15 \Omega$ (for uses $\cap$, 75 or Dep. 21.25 and 9 other than Bass Guitar or $\mathbf{t} / \mathrm{J}$ mithly payme

## GROUP IISCO EUUPMENT

 F.A.L. PHASE 50 MK.III AMPLIFIER Terms: Deposit 86.50 and 9 monthly Termi: Depositpayment of 24.92 (Total 250.78 ) F.A.L. PEASE 60 MK.III AMPLIFIER PR. FANE POP 50 L/BPEAKERS Terms: Deposit 110 and 9 monthly payments of 85.37 (Total 858.33 )
F.A.L. PHABE 50 ME.II AMPLIFIER PAIR L12/25 25W L/S in cabints
Terms: Depoalt 210 and 9 monthly Terms: Depoalt 810 and 9 month)
parments of 85.88 (Total 882.82 ) LINEAR T40/60D AMPLIFIER PAIR L12/25 L/S in cabinets Terments of 38.56 (Total 267 month FAL PEASE 100 a MPLIPIEP 4 FANE POP 50 L/SPEAEERS Terms: Deposth 15.95 and 9 monthly payments of $810-73$ (Total $2118 \cdot 52$ )

## PACKAGE OFFERS

 \&34.95 PACKAGE PRICE enc $652{ }^{\text {atim}}$

PACKAGE PRICE 656 PACKAGE PRICE 659.95 PACKAGE PRICE £99.95 ${ }_{\text {ki }}^{\text {cam }}$

HIGH QUALITY LOUDSPEAKER UNITS ALL TWO TONE REXINE AND VYNAIR FINISH L125 50 WATT Fitted pair of $12^{\prime \prime} 50$ watt high flux speakers for conservative rating. Impedance $8-15$ ohms. Carr. $£ 1.50$
Or deposit $£ 4 \cdot 50$ and 9 monthly pavts. of $£ 3 \cdot 70$. 12/25 120 L12/25 12" 25 WATT L13 13" $\times \mathbf{8}^{\prime \prime} 10$ Watt 10,000 lines $\cap 11.0$. 10,000 lines 3 or 15 $\left.\begin{array}{l}15 \text { ohms. } \\ \text { Carr. } 50 \text { p. }\end{array}\right\} \begin{aligned} & \text { ohms. State imped. } \\ & \text { ance required. Carr } 40 p\end{aligned}$

RSC BASS-REGENT

50 WATT AMPLIFIER A powerful high quality all-purpose unit for lead, rhythm, bass guitar, vocalists, gram, radio, tape. Peak
Output rating. Loudspeaker unit Output rating. Loudspeaker unit
optional horizontal or vertical mounting

* Two extra heavy duty 12 in . 50 m L'sples. Wour Jack inputs and two Volume Controls for instant use of up to four pick-ul
or "mikes". Bass and Treble controls. Credit Terms: Deposit Treble controls. Credit Terma: Depoalt
815 \& 9 monthly
DYts of 86.25 Carr. St A.E. pyta of 88.25
(Total 871.25 )
RSC GP30 HI-FI AMPLIFIER


For Guitar, Vocal or Instrumental Group A 4 input, 2 vol. control Hi-Fi 30 watt unit with separate Bass and Treble controls.
Current valvps. Peak output rating. Strong Rexine covered cabinet with handles. Atrac-
tive black/gold P.Y.C. facia. For 3 or 15 ohm speakers, Bend \&.A.E. for fin $\$ 7$ Carr. leaflet. Terıns: Deposit $£ 4.40$ anil 9 monthly th $\iint_{65 p}^{\text {Ca }}$ payments of 82.45 (Total 296.45 )

monthly $\mathrm{L} G U \mathrm{~J}_{65 \mathrm{p}}^{\text {Car }}$

CREDIT TERMS AVAILABLE ON PURCHASES OVER 68 (Kits Excepted) INTEREST CHARGES REFUNDED ON CREDIT SALES Settled in 3 months
NOW THREE NEW BRANCHES 19/19A MARKET STREET NOTTINGHAM ALSO 5 MARKET SQUARE SUNDERLAND 8 LITTLE UNDERBANK STOCKPORT
AND NEW MUCH LARGER PREMISES 24 NEWGATE CENTRE, NEWCASTLE


MAIL ORDERS \& EXPORT ENQUIRIES TO
AUDIO EOURE, HENCONNER LANE, LEEDS, 13. Tel: Pudsey (09735) '77631.
PERMS C.W.O. or C.O.D. No C.O.D. under 21. OVER 2 OR AS QUOTED. TRADE SUPPLIED SA E. PLEASE WITH ENQUIRIES. SUPPLIERS TO GOVERNMENT DEPARTMENTS EDUCATION AOTHORITIES, SCHOOLS, HOSPITALS, H.M. FORCBS ETC. MAIL ORDERS MOST NOT BE SENT TO SHOPS
R.S.C. Branches listed below open all day Saturdays and operate a 5 -day week BRADFORD 10 North Parade (Closed Wed.). Tel. 25349 BLACKPOOL (Agent) O \& C Electronics BIRMINGHAM $30 / 31$ Great Western Arcade,

Tel. $021-2361279$ (Closed Wed.) DERBY 26 Osmaston Rd., The Spot (Closed Wed.), DARLINGTON 18 Priestgate (Closed Wed) Tel 68043 EDINBURGH
101 Lothian Rd. (Closed Wed.). Tel. 2299501 HUL Paran Stel. 2484158
 LEEDS 5-7 County (Mecca) Arcade, Briggate (Closed Wed.). Tel. 28252 LIVERPOOL 73 Dale Street (Closed Wed.) Tell. 2363573 LONDON 238 Edgware Road, W. 2. (Closed Thurs.) ${ }_{\text {Tel. }}$. 23 i 269 MANCHESTER G0A OIdham Street, (Closed Wed.) $\begin{gathered}\text { Tel. } \\ \text { Th } \\ 2778\end{gathered}$ miodlesbrough to6 Newport Rd. (Closed Wed).

## NEWCASTLE UPON TYNE

24 Newnate Shopping Centre (Closed Wed.). Tel. 21469 NOTTHG GAM 19/19A Market Street (Closed Thurs.) SHEFFIELD 13 Exchange Street (Castle Market Blds.) SUNDERLAND 5 Market Sq. (Closed Wed.) Tel. 70573

## concal PEAGA of the yegr's meseraing - Alonion

Join thousands of Hi-Fi enthusiasts in an atmosphere that will set your ears tingling with excitement. Whatever your taste classics, jazz or pop - there will be something for you. The international giants of the Hi-Fi industry will be there - showing the latest specialist equipment: unit audio; tape recorders, cassette, cartridge and reel-to-reel; loudspeakers; AM/FM tuners, receivers; amplifiers, radios; tapes and discs. There will be continuous demonstrations in over 65 special Audio Studios - plus daily presentations, lectures and discussions in the Hi-Fi Theatre. And the Sunday Mirror is offering concession rates of admission to its readers, plus exciting competitions. Don't miss this great event.


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Tuesday 24 October to
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Admission 40p



Burns and scalds can rob your fingers of the dexterity essential to your hobby. So keep Burneze close to hand for minor burns. This unique scientific aerosol cools the heat out in just 8 seconds, anæsthetizes pain, reduces swelling. Be well-equipped-buy Burneze, from chemists.


Potter \& Clarke Ltd Croydon CR9 3LP

only 80 p

M. MOLE \& SON LIMITED, NEWPORT, MON. NPT 5XZ.


The leakage current of the NEW $\times 25$ is only a few microamps and cannot harm the most delicate equipment even when soldered "Iive" Tested at $1500 v$. A.C. This 25 watt iron with it's truly remarkable neat-capacity will easily "out-solder" any conventionally made 40 and 60 watt soldering irons due to its unique construction advantages.

Fitted long-life iron-coated bit $1 / 8^{\prime \prime}$
The 15 watt miniature model CCN also has negligible leakage Test voltage 4000 v . A.C Totally enclosed element in ceramic shaft. Fitted long-life iron-coated bit $3 / 32$ 4 other bits available $1 / 8^{\prime \prime}, 3 / 16^{\prime \prime} 1 / 4^{\prime \prime}$ and $1 / 16^{\prime \prime}$ PRICE: $£ 1.80$ (rec. retall) 2 other bits available $3 / 32^{\prime \prime}$ and $3 / 16^{\prime \prime}$.

OR Fitted with triple-coated, (iron, nickel and Chromium) bit $1 / 8^{\prime \prime}$
PRICE: $£ 1.95$ (rec. retail)

Totally enclosed element in ceramic and steel shaft Bits do not "freeze" and can easily be removed

PRICE: $£ 1.75$ (rec. retail) Suitable for production work and as a general purpose iron

## A SELECTION OF OTHER SOLDERING EQUIPMENT.



## MODEL CN

Miniature 15 watt soldering iron fitted $3 / 32^{\prime \prime}$ ironcoated bit. Many other bits available from $3 / 64^{\prime \prime}$ to $3 / 16^{\prime \prime}$. Voltages $240,220,110,50$ or 24
PRICE: $£ 1.70$ (rec. retail)
MODEL CN2
Miniature 15 watt soldering iron fitted with nickel plated bit $3 / 32^{\prime \prime}$. Voltages 240 or 220 PRICE. £1.70 (rec. retail)


MODEL G
18 Watt miniature iron, fitted with long life iron coated bit $3 / 32^{\prime \prime}$. Voltages 240,220 or 110 . PRICE. $£ 1.83$ (rec. retail)

contains 15 Watt miniature iron fitted with $3 / 16^{\prime \prime}$ bit, 2 spare bits $5 / 32^{\prime \prime}$ and $3 / 32^{\prime \prime}$, heat sink, solder, stand and "How to Solder" booklet. PRICE £2.75 (Rec. retail)

MODEL SK. 2 KIT
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[^1]
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## IP <br> H-P. (Electronics) Ltd

## THE HY41

The HY41 supersedes the popular HY40 introduced by ILP last year. This highlv improved module achieves true High Fidelity with a dramatic reduction in distortion (typically 0.05\% at 1 KHz into 8 ohms!) and is electronically and mechanically compatible with the HY 40 .

With this important improvement the HY41 retains all of the quality characteristics found in the earlier version and PC. board Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete nower amplifier of extremely high pertormance sufficiently versatile to provide power not merely for Hi-FI but also for nublic address systems and industry.

The free mamual gives a full circuit diagram of the HY41 and its various applications including a complete stereo amplifier

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over lecent years
OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts
R.M.S. continuous

LOAD IMPEDANCE: 4-16 ohms.
INPUT IMPEDANCE: 30 K ohms at 1 KHz
VOLTAGE GAIN: 30db at 1 KHz
TOTAL HARMONIC DISTORTION: less than 0.15\% (typical 0.05\%)
at 1 KHz .
FREQUENCY RESPONSE: $5 \mathrm{~Hz}-50 \mathrm{KHz}+1 \mathrm{db}$
SUPPLY VOLTAGE: +22.5 volts D.C.
SUPPLY CURRENT: $\overline{0} .8 \mathrm{amps}$ maximum.
PlilCE: inc. comprehensive mantal, P.C board, five extra components and $P$. \& $P$.
MONO: $£ 4.90$
STEREO: 99.80

## UNIQUE HYBRID PRE-AMPLIFIER

The HY5 has rapidly established a position in the WORLD as the sole hybrit pre-amplifier to contain all feedback and equalization networks within an integrated preamplifier circuit

Supplied with the HY5 are two stabilizing capacitors and by the addition of volume, treble and bass potentiometers it is ready for use.

Internally the HY5 provides equalization for almost every concelvable input, the desired function is achieved by use of a multi-way switch or by direct interconnection.

Two distinctive features of the HY5 are its inbuitr stabilization circuit, allowing it to be run off any unregulated power supply from $16-25$ Voles and a balance circuit which, when linked by a balance control to a second HY5, lorms a complete stereo preamplifier.

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely fow noise with a high overioad capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system

INPUTS
Magnetic Pick-up (within +1 db RlAA curve) $2 \mathrm{mV} .47 \mathrm{~K} \Omega$
Tape Replay lexternal components to suit head). $4 \mathrm{mV} .47 \mathrm{~K} \Omega$
Microphone (flat) $10 \mathrm{mV} .47 \mathrm{~K} \Omega$
Ceramic Pick-up fequalized and compen-
satable) 20-2000mV. variable
Tuner (flat) $250 \mathrm{mV} .100 \mathrm{~K} \Omega$
Auxiliary $1250 \mathrm{mV} .47 \mathrm{~K} \Omega$
Auxiliary $22-20 \mathrm{mV} .100 \mathrm{~K} \Omega$

OUTPUTS
Main Pre-amp output 500 mV Direct tape output 120 mV

ACTIVE TONE CONTROLS (Bexendall)
Treble $\pm 12 \mathrm{db}$
Bass $+{ }^{-1} 12 \mathrm{db}$
INTERNAL STABILIZATION
Enables the HY5 to share an unregulated supply with the Power Amplifier.
SUPPLY VOLTAGE
$16-25$ volts
PRICE: MONO: $£ 3.60$ STEREO: $£ 7.20$


SUPPLY CURRENT 6mA approx.
OVERLOAD CAPABILITY
better than 26 db on most sensitive input infinite on tuner and auxi. OUTPUT NOISE VOLTAGE: 0.5 mV

POWER SUPPLY PSU45
The versatile P.S.U. 45 is designed to supply your HY 41 's + HY5's in stereo or mono format.

Specification
Input: 200-240 Volts.
Output: $\pm 22.5$ Volts at 2 amps.
Overall Dimensions: L. $7^{\prime \prime}$; D. 3.8'; H. 3.1"
PRICE. $£ 4.50$ inc. P. \& P.

## TUNERS

Please add 75p P. \& P. Amstrad Multiplex 3000 Armstrong 523
Armstrong 524
Rogers Ravensbrook FET4 Rogers Rave
(Cased)
Rogers Ravensbourne FET. 4
(Chassis)
Rogers Rave
Rogers Ra)
(Cased)
Sinclair PRO60 (Module) Leak Detea FM (Cased) $\begin{array}{ll}\text { Sinclair 2000/3000 Tuner } & 631.95 \\ \text { Leak Delta FM (Cased) } & 65475 \\ \text { Leak Deita AM } / F M \text { (Cised) } & 664.50\end{array}$

## TUNER/AMPLIFIERS

Please add 75p for P. \& P Alpha Highgate 150
Armstrong 525 (Teak
Armstrong 526 AM/FM (Teak cased) Leak Delta 75 Philips RH781 Telecon 2100
Goodmans One Ten
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Rogers $R$ brook (Teak)
Rogers $R$ /brook (Chassis) Alpha FR 3000

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Sinclair PRO60 $2 \times \mathrm{Z} 30 / \mathrm{PZ} 5$ Sinclair PRO60 $2 \times 230 /$ PZ 6 Sinciair PRO60 $2 \times$ Z50/PZ8/Trans Sinclair AF
Sinclair 605
Sinclair 2000 Mk .11
Sinclair 3000 Mk
Sinclair 3000 Mk . 11
Wharfedale Linton
Goodmans Max Amp
Teleton SAQ30
Rotel RA 210
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$E 43.95$
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GOLDRING GL72 Chiss
GOLORING GL72P
GOLDRING GL85
GOLDRING GL85P
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TANDBERG $3021 \times$ twin track stereo
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AKAI GXC 40D Tape Deck
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APCo Batt/Mains with AM/FM
Radio
BUSH Discassette DC70
BUSH TP 66 Batt $/ \mathrm{Mains}$
BUSH TP 70 battery/malns
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DECCA 2000 (batt./mains)
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JVC NIVICO 9420 L/S with AM/FM
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PYE 9118 Stereo
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DJ. 30 L ( 3000 w ) 3-channel paychedelic light DJ. 30L (3000w) 3-channel paychedelic light unit is a standard itting. Deck cut out switches $38^{* *} \times 27^{*} \times 10^{*} .8224 .00$ Buper plum a third turntable which can be used for Jingles or other effects without using the main deck nystem. Flexi lights are alvo Ktted. Bize $50^{\prime \prime} \times 27^{\prime \prime} \times 10^{\circ}$. $2261 \cdot 00$.

## PA-DISCO AMPLIFIERS



DISCO-AMP. 100 watt rms. output for $8-16$ ohms, 4 channel inputs, 2 -mic, 2 decks. Separate volume control plus masters. Response 30 Hz -30 KHz , distortion less than $1 \%$. Treble/Bass/ PFL/Mic over-ride etc. Panel size $16 \frac{1}{2}^{\prime \prime} \times 7^{\prime \prime}$.

DJ. 70 S MIXER/AMPLIFIER. 70 watt rms. output for 8-16 obms. 2-mic, 2-aux/decks. Master volume/Bass/ Treble. Size $15 \frac{1^{\prime \prime}}{} \times$
DJ.105S. 30 watt rms. version. Size $113^{\prime \prime} \times 5^{\prime \prime} \times 6^{\prime \prime}$. DISCMASTER SLAVE AMPLIFIER. 100 watt rms. for 8 - 16 ohms. $£ 59 \cdot 50$

NEW D.J. 500 SERIES P.A. AMPLIFIERS 50 WATT, 70 WATT \& 100 WATT MODELS This new range incorporates many features that make them ideal for the protessional user. clubs, tiscotheques, factories ete. Fibre glass P.C. Boards are used throughout with low noise silicon transistors, high stability resistors, generously rated components and hand w
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* Built in hasa boor
below 30 Hz .
* 4 channel mixer with slider contruls.

All three amplifers have a built in emitter follower output socket for connecting a slave amplifier to enable the power output to be increased amplifers and a separate matching 100 v line transformer is available sPECIFICATION
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Harmoric Distortion less than $1 \%$
Speaker Impedance $8-16$ ohms.
位 (all 00 mV at 1 meg ohnt.
size (all models) $15 \mathbf{z}^{\circ} \times 5^{*} \times 6^{\circ}$.
Power Output: Model T.J. 500
50 watte R.M.s. $\quad 856.25$ Model D.J. $700 \quad$ - $\quad 70$ watte R.M.S. $\quad 867.50$ Model D.J. $1000 \quad-\quad 100$ watis R.M.S. $279 \cdot 00$

## DISCO MINI

Hardly larger than a suitcase yet contains all the necessary realures bile undt. The pre-amp bile unct. The preamp trols tor both mic. antl decks
 vidual volume controls and inputs.
plus the addition of a cross fade for deck to deck sound transfer. A built in P.F.L. system for cueing, together with mic-over-ride facility are standard on all units. Response $\mathbf{2 0 - 2 0 , 0 0 0} \mathbf{H z}$. Mic. input 5 mV . McDonald M.P. 60 . McDonald M.P. 60 Turntables are used with high quality ceramie cartridges, and each deck has its own individual cut out switch fitted. for use with the - Disenaster' 100 watt power amplifier as above. for use with the 'Disemaster'
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DISCO COLT. 150 watt. LIQUIMATIC MINI, 50 watt Q1 with $6^{\circ}$ wheel DISCOWHEEL, 50 watt Q1 with quick change GNOME 150, 150 watt Q1 with Cassette LIQURMATIC, 150 watt $Q 1$ with 6 wheel PLUTO TUTOR-2, 250 watt Q1 with Casse and $6^{\circ}$ wheel.
TUTOR-2, with Llquisplode TankLiquid Wheel and Crystal Wheel Liquill Liquid Wheel and Crystal Wheel Liquitd
Cassette and Moire ( 24 different types $t$, choose from). Portable Hi-Power Strobes YOU WILL BE SURPRISED BY THE IDY PPICE O DनPTOPMANCF

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## DJ.101. Battery powered, fochannel. variable

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## THIS IS THE FIRST PAGE OF THE GREAT Bl-PAK SECTION

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BA154
BA155
BA156
BY100
RY101
BY 105
BY114
BY126
BY127
BY128  $\begin{array}{ll}14 & \text { CG651 } \\ 0 & \text { (Eq) OA70 }\end{array}$




0.08



DIODES AND RECTIPIERS

2 N 305
$2 N 305$
$2 N 339$ $\begin{array}{r}3054 \\ N 3055 \\ \hline\end{array}$

2N4059
$2 N 4060$
$2 N 4061$
$2 N 4062$
$2 N 4284$
$2 N 4285$
$2 N 4286$
$2 N 4287$
$2 N 4288$
$2 N 4289$
$2 N 4290$
$2 N 4291$
$2 N 4292$
$2 N 4293$
$2 N 5172$
$2 N 5457$
$2 N 5458$
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28322 A


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C13 20 Electrolyties Trans. types
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$\begin{array}{lr}C 17 & 10 \\ \text { C18 } & 4 \text { Rorted Control Knobs } \\ \text { Rotary Wave Change } 8 \text { wit }\end{array}$
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U10 20 BAY50 charge atorage Dlodes DO-7 (Glars
U11 25 PNP All. Planar Trans. To.5 Hke 2N 1132, 2 N2904
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Ul5 25 NPN Sill. Planar Trana, TO-5 like BFY51, 2 N697 U16 103 Amp Sillicon Rectiders Stud Type up to 1000 PI U17 30 Germanium PNP AF Transistors TO-5 like ACY 17.22 $\mathrm{U} 18 \quad 86 \mathrm{Amp}$ gilicon Rectiflers BYZ13 Type up to 600 PIV .. 0.50 U19 25 8liticon NPN Transiatora like BCl08

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| :--- | :--- | :--- |
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 U32 25 Zener Dloder 400 mW D $0-7$ care $3-18$ volts mixed
U33 15 Plastic Case 1 Amp Sllicon Rectifers IN4000 Beries
U34 30 8illcon PNP Alloy Trans. TO-5 BCY $2628302 /$
U35 25 8iltcon Planar Transistors PNP TO-18 2 N2906
U36 25 8ilicon Planar NPN Transistors TO-5 BFY50/51/5
U37 30 Eillcon Alloy Transistors 80-2 PNP OC200, 28322.
U38 20 Fast Awitching Silicon Trana. NPN 400 MHz 2 N 3011
U39 30 RF. Germ. PNP Tranaistora $2 \mathrm{~N} 1303 / 5$ TO-5
U40 10 Dual Transiniora 6 lead TO-5 2N2060

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| T43 | 25 | Bil. Trans. Plastle TO-18 A.F. BCl13/114 |
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| :---: | :---: | :---: | :---: |
| Anode voltage (Vdc) | 170min | 175 maln | 5 |
| Cathode Current (mA) | 23 | 14 | 8 |
| Numerical Height (mm) | 16 | 13 | 9 |
| Tube Helght (mm) | 47 | 32 | 22 |
| Tube Diameter (mm) | 19 | 13 | 12 wide |
| 1.C. Driver Rec. | $\begin{gathered} \text { BP41 or } \\ 141 \end{gathered}$ | $\begin{gathered} \text { BP41 } \\ 141 \end{gathered}$ | BP47 |
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Built to a speciflcation and NOT a price, and yet still the greategt value on the market the PAl00 stereo pre-amplifier has been conceived from the latest circuit teconiques. Destgned for use with the AL50 power ampliffer system, this quality made unit incorporates no less than eight sificon planar transistor
Three gwiltcheil stereo inputs, and rumble and scratch fitters are features of the PA100 Three switcheil stereo inputs, and rumbe switch, volume, balance and continuously variable bass and treble controls. SPECIFICATION Frequency Response inputs: 1. Tape Head
2. Radio, Tuner
3. Magnetic P.U.
$20 \mathrm{~Hz}-20 \mathrm{KHz} \pm 1 \mathrm{~dB}$
better than $0 \cdot 1 \%$.
1.25 mV into $50 \mathrm{~K} \Omega$

35 mV into $50 \mathrm{~K} \Omega$
1.5 mV into $50 \mathrm{~K} \Omega$

All input voltages are for an output of 250 mV . Tape and P.U. Inputs equalised to IRLAA curve within $\pm$ lab. from 201 Iz to 20 KHz .
Basa Control
$\begin{array}{ll}\text { Filters: Rumble (IIgh Pass) } & \mathbf{1 0 0 H z} \\ \text { Beratch (Low Pass) } & \mathbf{8 K Y}\end{array}$
Signal/Noratch (Low Pass)
Input overloal Ratio
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Supply
Dimerations
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MULTI-CHANNELLEO SOLDER


Pre fluxed solder 'READY TO USE' produced to B.S.I. 441 IN HANDY SIZES f0. 41 and f. $1 \cdot 16$

## Is Ignorance Bliss?

MORE than $£ 100,000,000$ 's-worth of audio equipment, half of which is hi-fi, is being sold in the UK every year. But according to the findings of the commercial research department of Philips Electrical, who published details of a survey earlier this year, this buoyant industry thrives in a sea of appalling ignorance.

The report doesn't mince words. It says that 'the level of ignorance encountered in a number of shops was startling". Many of them had inadequate audio facilities and expertise. When questioned, most dealers considered that the most important factor was the price, this being considered far ahead of design and performance.

This will come as no surprise to knowledgeable types who have had to buy goods from such shops. But the report goes on to say that the customers' largest single source of information is not from friends, magazines, sales literature or advertisements, but from dealers. It goes without saying, then, that in this classic example of the blind leading the blind, the average customer comes out of the ordeal in a state of mental fog.

To quote from the results of random samples: $44 \%$ did not know whether their cartridge was magnetic or ceramic; $75 \%$ did not know the frequency response of their amplifier; $60 \%$ could not give the audio output, and of those who could, $33 \%$ did not know if the figure was r.m.s. or music rating. Of more pertinent interest is that many people do not know whether or not they are listening to stereo. Many thought they were hearing stereo on Radios 1 and 2 and $75 \%$ of intending buyers interviewed thought they heard stereo on Radio 1!

It does not need a research team to expose the sad lack of competence (and, often, interest) in many shops, and it follows logically that many customers must be similarly ill-informed. However, we feel that there is a dividing line between the depth of knowledge a customer needs to use his equipment properly and the knowledge which is simply a technical luxury. In view of this, we were delighted to read that, despite the epidemic of ignorance, $64 \%$ of customers were "very satisfied" with their equipment.

One trade journal has suggested that we want a "big effort to create more unsatisfied customers". However, although we want people to appreciate good sound reproduction, is it ethical to unsettle people who are perfectly content with their equipment in order to gain a few more kHz and a little less distortion which they probably could not distinguish? In general we are against ignorance but sometimes, especially in audio, it can be bliss!

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[^3]
# NEWS... 

## Mini calculator

Sinclair Radionics Ltd., annouce the world's smallest electronic calculator, just over ${ }_{4}{ }_{4} \mathrm{in}$. thick,

The Sinclair "Executive" performs all the functions of large desk machines although it is only 2 inches wide by $5^{1}{ }_{2}$ inches long. The total weight, including batteries, is $2^{1} 2$ ounces. The illuminated display has a capacity of 8 digits and the machine will add, subtract, divide and multiply virtually instantaneously. Other features include automatic squaring, reciprocals, fixed or floating decimal point operation, and a memory for locking in instructions to repeatedly multiply or divide by a pre-determined factor.

This calculator uses a single MOS chip containing 7,000 transistors. Normally this type of I.C. consumes in the region of 350 mW which alone would rule out a really slim pocket calculator because of the large batteries that would be required. The Sinclair calculator, however, includes a special circuit (patents applied for) which reduces the consumption to an average of only about 20 mW . This is achieved in two ways; firstly by switching off the I.C. completely in between clock pulses, relying on the capacitance associated with the active elements to store the information, and secondly by reducing the clock rate considerably (and hence the "on" period) when the calculator is not actually performing a calculation.

Another factor contributing to the low overall power of the calculator is the way in which the display is driven. The battery voltage is chosen so that it is just sufficient to drive the monolithic gallium arsenide phosphide display directly so that no power is wasted in large dropper resistors. Indeed no such resistors are used at all as the display is driven directly from the chip. The batteries themselves are three tiny mercury cells of the type commonly used in hearing aids. These provide 20 hours of life.

The keyboard, another item which contributes to the thick-

ness of competitive machines, was specially developed for this calculator and is extremely thin, being only 4 mm iu depth. Price is $£ 79$ and every unit is guaranteed five years. Sinclair Radionics Limited, London Road, St. Ives, Huntingdonshire PE 17 4HJ. St. Ives (04806) 4311.

## Texan Reprints

Reprints of the "Texan" $20+20 \mathrm{~W}$ Stereo Amplifier are now available.

They may be obtained by sending $35 p(30 p+5 p$ postage/packing) to Practical Wireless Editorial, Fleetway House, Farringdon Street, London, EC4A $4 A D$. - Hurry the demand will be great!

## Sheffield exhibition

Sheffield Education Committee will be holding an exhibition next month with the theme "Communications Today" at the Yewlands Comprehensive School. Creswick I ane, Grenoside, Sheffield, S30 3NN.

The dates of the exhibition are Friday and Saturday 3 rd and 4th November and further information may be obtained from J. M. Nicholson, M.A., at the Yewlands School, telephone Ecclesfield 3778 ( f 'om London the code is 07415 3778)

## Record care

New from Bib is the Record Care Kit Ref. 43. Contained in a presentation 3 -colour printed see-through box will be a new all-chrome finish Groov-Kleen Automatic Record Cleaner incorporating counterweight, arm-rest and all the features of the other two Bib Groov-Kleen's Models 42 and 50, turntable spirit level, plastic record dust-off, stylus \& turntable cleaning liquid, right angle cleaning brush suitable for stylus and Groov-Kleen pad, and a turntable cleaning cloth. Recommended retail price for the Record Care Kit Ref, 43 is $£ 2 \cdot 49$, including P.T.


## NEWS...

## Siemens equipment

The first Siemens record player to become available in the UK is announced by Interconti Electronics, who also announce the introduction of the RS303 tuner amplifier.

The RW346 stereo record player, comes complete with plinth and a protective dust cover. The arm is fitted with Shure magnetic pickup system and diamond stylus. The cabinet is available either in white or a light satinwood finish.

The recommended retail price of the unit is $£ 100 \cdot 66$ p.

The RS303 tuner amplifier to which the RW346 can be linked, is a stereo drive unit providing quadrosound reproduction. The receiver is equipped with built-in antennae covering all wave ranges. Station selection in the f.m. range is accomplished easily by using the five pre-set tuners each of which is tuned by a given control button. Price is $£ 180$. Interconti Electronics Ltd., Albany House, Petty France, London, S.W.1.

## Strobe unit

Dawe Instruments Limited announce stroboscope, Type



## R.F. signal generator

Taylor Electrical Instruments have introduced the re-styled 68A/M Mk 11 signal generator which covers 100 kHz to 240 MHz in eight switched ranges with $30 \%$ internal amplitude modulation between 50 Hz and 5 kHz . All controls and sockets are panel mounted and include Set r.f. Level (with meter indication), Set a.f. Level, r.f. Output (switched 1-10). Attenuator ( 0 to -80 dB .).

1209D. In addition to the many features of the well established Transistor Stroboflash, the new design provides a high sensitivity input (minimum external trigger: 200 mV r.m.s.) which enables the Stroboflash to be triggered directly from such sources as vibration meters, vibration analysers or from an electromagnetic pickup.

The Stroboflash produces a high iutensity white light and is capable of directly measuring the speed of any shaft or rotating part withir the range $300 \mathrm{rev} / \mathrm{min}$ to $18,000 \mathrm{rev} / \mathrm{min}$. The very short duration of the light flashes makes it possible to measure speeds up to $180,000 \mathrm{rev} / \mathrm{min}$ ! Dawe Instruments Ltd., Western Avenue, London, W3 OSD.

## Stubec Radio

The above company have published a new catalogue which contains details and prices of radio components and special details of do-it-yourself loudspeaker components and grill fabrics.

For your copy of this catalogue, send a stamped addressed foolscap envelope to: Stubec Audio, 2 St. Faith's Street, Maidstone, Kent.


The Super Sovereign RP75 covers medium, long, two short and f.m. bands and it's the latest in the Hacker range of portable radios, taking over from the well-known Hacker "Sovereign." Retail price is $£ 69$.

Practical Wireless will be on Stand 18 (ground floor) at the International Audio Festival and Fair, Olympia from October 23rd-28th.

One of the highlights of our stand will be a display devoted to the "TEXAN" $20+20 \mathrm{~W}$ Stereo Amplifier. You will also be able to get a reprint of the "TEXAN" for 30p.

Amongst other items featured, there will be a Reverberation Unit, Loud-Hailer, Buzz-Bar, Car Radio, Miniature Organ, Radio Transmitter, Receiver, Power Supply and Audio Test Instruments.


## PART ONE

Reverberation is an acoustical effect caused by the reflection of sound from hard surfaces within an enclosed space such as a large room or hall containing little sound absorbing material such as carpeting and padded furniture etc. The reflections become multiplied as the sound bounces from surface to surface and reaches the ears as a series of echoes of very short duration. The overall effect however, is a sustained sound because the reflections may take a considerable time to die away completely, depending on the volume of the enclosed space. In a very large hall the dying away, or decay time, may be as much as several seconds. Reverberation is quite unlike the more distinct single echoes caused by sound reflection in the open air from a single surface, such as a large building, or the echo effect produced by feeding a recorded signal from tape back into the recording amplifier.

## Spring Line Reverberation

Reverberation is quite easy to produce artificially. All that is required is a) a means of obtaining a series of echoes of short duration, of a few milliseconds in fact, and b) a means of sustaining repetition of the echoes for at least two seconds. A length of coiled springy wire will do this if it is mechanically excited at one end; this can be done by using a simple electromagnetic sound transducer. The speed of sound vibration up and down the spring line is very slow when compared with the speed of elec-

## Specification

## Microphone Sensitivity

Line Input Sensitivity
Microphone Input Impedance
Line Input Impedance Output Impedance
Frequency Response (without reverb)
Signal to Noise (without reverb)
Distortion Factor (without reverb) Reverb Delay Time

0.5 mV for 800 mV output (without reverb) 200 mV for 800 mV output (without reverb) Nominal 100k $\Omega$<br>Nominal $10 \mathrm{k} \Omega$<br>Nominal $5 \mathrm{k} \Omega$<br>20 Hz to 20 kHz , either input<br>Microphone: -60 dB<br>Line: - 80 dB<br>Less than 0.5 per cent

Approx 2 seconds maximum
trical sound through wires. The spring line offers little resistance to vibration travelling along it and because of this the vibrations will continue to travel up and down the line for some time. The strength of the vibrations do, however, decrease slowly and eventually cease altogether.
 reverberation unit.


Fig. 2: Circuit diagram of the complete reverberation unit.

## The complete reverberation unit

The block diagram in Fig. 1 will help to clarify the principle of a spring line reverb system. First the direct signals from a microphone or other external signal source are taken into the appropriate input. These are amplified and taken both to the output amplifier and to the spring line driving amplifier. The output from the latter drives the spring line and the reverberated signals are taken from the other end of the line, amplified and mixed with the direct signal. The output signal is therefore a mixture of both direct and delay signals. A control is inserted so that the level of signal from the reverberation circuit can be adjusted to provide different reverberation levels or it can be cut off altogether. One can simulate the short reverberation of a small room or the long sustained reverberation of a large hall.


Fig. 3 : Frequency response of the spring line driver ampllfier.

## The Circuit

The complete circuit is given in Fig. 2. The microphone preamplifier $\operatorname{Tr} 1$ and $\operatorname{Tr} 2$ is suitable for microphones of $200-600 \Omega$, or low impedance microphones with a matching transformer, or with a crystal microphone. The output from this is taken to a passive mixing network via the microphone gain control VR1. The line level signal input, suitable for any linear signals from about 200 mV upward, is taken via the line input gain control VR2 and thence to the passive mixing network R7 and R8.

The mixing network losses are recovered by the amplifier $\operatorname{Tr} 3$ and the output from this is taken via a passive network to a) the output emitter follower stage $\operatorname{Tr} 4$ and $b$ ) to the reverb unit driving amplifier $\operatorname{Tr} 5, \operatorname{Tr} 6, \operatorname{Tr} 7$ and $\operatorname{Tr} 8$. This amplifier delivers around 1W at full reverb i.e., with the reverb control VR3 at maximum. The amplifier also has a special frequency response as shown in Fig. 3 to suit the response require-


Fig. 4 : Frequency response curve of the reverb output amplifier.

(a)

(b)


Fig. 5: Ways of using the spring line reverb unit.


Fig. 6: Details of the front panel component positions and the drilling dimensions
ments of the spring line driving transducer.
The output from the driving amplifier is taken to the spring line unit socket marked 'Input' and also to the reverb overload meter circuit R37, D1, C20 and the $50 \mu \mathrm{~A}$ meter. The use of the meter will be explained later.

The output from the spring line unit, from the socket marked 'Output', is taken to the amplifier $\operatorname{Tr} 9$ which also has a special frequency response as shown in Fig. 4. The signal from this amplifier is then taken to the emitter follower output stage Tr4.

The output from $\operatorname{Tr} 4$ is split to two output sockets, both of which can be used to feed a tape recorder and/or amplifier, or two tape recorders, or amplifiers as required and as shown in Fig. 5.

The reverb unit is mains operated by the trans-
former T1, bridge rectifier BRI and smoothing network C13, R 23 , and C 12 . The supply rail at Cl 3 is 18 V positive and this goes direct to the spring line driving amplifier. Further smoothing is needed for the signal amplifier stages, hence the addition of R 23 and C 12 from which the supply rail is 15 V positive. Note that the transformer MT280 has two 6 V secondary windings which are connected in series to provide 12 V to the bridge rectifier $B R 1$.

## Construction

The spring line reverb unit is a type HR162 (Henry's Radio Limited) and is $16^{7} 3_{3}$ in long. This will just fit into a Lektrokit case (17in long) in which the prototype unit as shown in the photos was built

## components list

| esis | rs |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R1 | 680, 2 | R16 | 1 k , | R31 | 680, |
| R2 | 150k $\Omega$ | R17 | 22k $\Omega$ | R32 | $56 \Omega$ |
| R3 | 150k』2 | R18 | 10 k ת | R33 | $1 \mathrm{k} \Omega$ |
| R4 | 390kS | R19 | 4.7ks | R34 | $0 \cdot 5 \Omega$ |
| R5 | 1 kS ) | R20 | $1 \mathrm{k} \Omega$ | R35 | $0 \cdot 5 \Omega$ |
| R6 | 10ks | R21 | 4.7kS | R36 | $56 \Omega$ |
| R7 | 15kS | R22 | 4.7kS | R37 | 22k $\Omega$ |
| R8 | 15k $\Omega$ | R23 | 680, 2 | R38 | 12kS2 |
| R9 | 12kS | R24 | 47kS | R39 | 120 k , |
| R10 | $120 \mathrm{k} \Omega$ | R25 | $15 \mathrm{k} \Omega$ | R40 | $1 \mathrm{k} \Omega$ |
| R11 | 10kS | R26 | 120ks | R41 | 220, |
| R12 | 1kS | R27 | 39k | R42 | 10ks |
| R13 | 100S | R28 | 22, | R43 | $1 \mathrm{k} \Omega$ |
| R14 | $56 \mathrm{k} \Omega$ | R29 | $47 k \Omega$ | R44 | $68 \mathrm{k} \Omega$ |
| R15 | $10 \mathrm{k} \Omega$ | R30 | 390S |  |  |
| All | esistors | 5 p | $r$ cent | ce | xcept | and R35 which are 1 W .

Capacitors

| Capacitors |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| C1 | $5 \mu \mathrm{~F}$ | C 10 | $100 \mu \mathrm{~F}$ | C 19 | $100 \mu \mathrm{~F}$ |
| C 2 | 100 pF | C 11 | $100 \mu \mathrm{~F}$ | C 20 | $10 \mu \mathrm{~F}$ |
| C 3 | $10 \mu \mathrm{~F}$ | C 12 | $250 \mu \mathrm{~F}$ | C 21 | $20 \mu \mathrm{~F}$ |
| C 4 | $5 \mu \mathrm{~F}$ | C 13 | $2000 \mu \mathrm{~F}$ | C 22 | $5 \mu \mathrm{~F}$ |
| C 5 | $5 \mu \mathrm{~F}$ | C 14 | $0.05 \mu \mathrm{~F}$ | C 23 | $1 \mu \mathrm{~F}$ |
| C 6 | $5 \mu \mathrm{~F}$ | C 15 | 6800 pF | C 24 | $0 \cdot 1 \mu \mathrm{~F}$ |
| C 7 | $50 \mu \mathrm{~F}$ | C 16 | $25 \mu \mathrm{~F}$ | C 25 | $5 \mu \mathrm{~F}$ |
| C | $100 \mu \mathrm{~F}$ | C 17 | $250 \mu \mathrm{~F}$ | C 25 | 6800 pF |
| C 9 | $5 \mu \mathrm{~F}$ | C 18 | 4700 pF |  |  |

All values above $1 \mu \mathrm{~F}$ are electrolytic types and should have a working voltage of at least 25 V , except C 13 which should be at least 35 V . Values below $1 \mu \mathrm{~F}$ can be any type, Mylar, ceramic etc.

| Semiconductors |  |  |  |  |
| :---: | :--- | :--- | :--- | :--- |
| Tr1 | BC109 | Tr5 | BC108 | Tr9 BC109 |
| Tr2 | BC109 | Tr6 | AC128 | BR1 BY164 or |
| Tr3 | BC109 | Tr7 | AC176 | similar silicon |
| Tr4 | BC109 | Tr8 | AC128 | bridge rectifier <br>  |

Miscellaneous
VR1, VR2 and VR3 Each 10kS log. Slider types Eagle RR11
PR1 $\quad 100 \Omega$ preset carbon type
Spring Line Unit Type HR162 (Henry's Radio Ltd.)
Transformer T1 240 V primary, $6 \mathrm{~V}+6 \mathrm{~V}$ secondary, 280 mA (Eagle Type MT280) Panel type (prototype is Eagle RR30)
On-off switch SW1 Toggle type
Overload Meter M1 50رA A, Eagle type MR2P
Jack Sockets $\quad 5$-off required
Phono Plugs 2-off required
Case Lektrokit, $17 \times 5 \times 5$ in. Made up from Front panel (1) LK401, Side Plates (2) LK311, Top, Back and Bottom Plates (3) LK501.
Available from Home Radio
Circuit Board
Aluminium

Plain Veroboard, $0 \cdot 15$ in, matrix, see text
16s.w.g. (for heatsink and power supply chassis); $z \times i \operatorname{in}$, angle.


Fig. 9 (above): Siting of the spring line unit within the case.

On the left is shown a view of the prototype illustrating the siting of the component boards.


Fig. 10: Details of the heatsink for Tr7 and Tr8.

The circuit board containing all the amplifier stages is mounted behind the front panel as shown in Figs. 7 and 8. The power supply is assembled as a separate chassis and this is mounted as shown. The spring line unit is mounted at the bottom and rear of the case and raised slightly on stand-off pillars as shown in Fig. 9 and with the output and input sockets facing towards the front.

The next stage is the circuit board which is $10^{7}{ }_{8} \times$ 5 in . and is mounted on the front panel by means of $3_{8} \times 3_{8}$ in. aluminium angle and the stand-off pillars as in Fig. 7. Note that $\operatorname{Tr} 7$ and $\operatorname{Tr} 8$ must be mounted on a heatsink as in Fig. 10 which is made of 16 s.w.g. aluminium to the dimensions given.


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THIS portable receiver uses silicon n.p.n. epitaxial planar transistors in the first four stages, followed by a transformerless output stage employing germanium n.p.n. and p.n.p. complementary transistors. The circuit is relatively simple, while offering good sensitivity and adequate power output.

When completed, the receiver has three wave-bands-long waves, medium waves and bandspreading of the h.f. end of the m.w. band (also sometimes termed Luxembourg Bandspread). However, though the switching associated with the provision of such bands is not very complicated, it does seem that it is sometimes a source of difficulty or wiring errors
during building. So the receiver is first constructed with one band coverage, tuning m.w. only. When it has been found to give proper results in this form, the few additional items required to add bandspread tuning and l.w. coverage are incorporated. It would, in fact, be easy to limit coverage to l.w. and m.w. only, if the bandspread range is not required, or even to omit l.w. where this band is not wanted.

## CIRCUIT

Fig. 1 is the circuit as constructed for one waveband, the m.w. ferrite rod winding L1 tuned by VC1, and the oscillator coil L3 tuned by VC2. VC1/2 is


Fig: 1. Circuit of the basic receiver for normal medium wave reception. The additional bandswitching circuit is shown in Fig. 4.


AFig. 2: Plan view of components mounted on top of circuit board with details of the mounting for the ferrite rod aerial. NOTE: Speaker should be shown connected to centre and RIGHT-HAND pin.

FFig. 3: Wiring on reverse side of circuit board together with transistor lead-out connections.


CKG24
a ganged capacitor with the trimmers TCl and TC 2 incorporated.

L2 is the m.w. base coupling winding for the oscillator/mixer $\operatorname{Tr} 1 . \operatorname{Tr} 2$ and $\operatorname{Tr} 3$ are the IF amplifiers using BF195's. Diode D1 is the detector or demodulator which also provides a.g.c. bias for Tr2. The collector circuit of $\operatorname{Tr} 3$ is decoupled by C6 and R7. Because of the very high gain of these transistors, it may be necessary to take certain steps to avoid squegging or instability, as described later. This depends to some extent on individual transistor characteristics.

VRI is the volume control feeding $\operatorname{Tr} 4 . \operatorname{Tr} 5$ and Tr6 are used in a directly coupled driver-output arrangement which will be found to give good results and plenty of volume with a $15 \Omega$ or $18 \Omega$ speaker.

## CONSTRUCTION

The controls are mounted on a $5 \times 2 \mathrm{in}$. "universal chassis" flanged member, Fig. 2. The top flange is cut off, to clear TC1 and the variable capacitor. When mounting this item, with its spindle through a clearance hole, it is necessary to use very short bolts, or

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## * components list

| Resistors |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R1 | $18 \mathrm{k} \Omega$ | R7 | $390 \Omega$ | R13 | $47 \Omega$ |
| R2 | $15 \mathrm{k} \Omega$ | R8 | $27 \mathrm{k} \Omega$ | R14 | $1.5 \mathrm{k} \Omega$ |
| R3 | $2 \cdot 7 \mathrm{k} \Omega$ | R9 | $2 \cdot 2 \mathrm{k} \Omega$ | R15 | 2.2S |
| R4 | 47k $\Omega$ | R10 | 270k $\Omega$ | R16 | $2 \cdot 2 \Omega$ |
| R5 | $120 \mathrm{k} \Omega$ | R11 | $220 \mathrm{k} \Omega$ |  |  |
| R6 | $330 \mathrm{k} \Omega$ | R12 | $680 \Omega$ | All | W 5\% |

VR1 $10 \mathrm{k} \Omega \log$. pot. with switch (S1)
Capacitors

| C1 $0.01 \mu \mathrm{~F}$ | C9 $0.01 \mu \mathrm{~F}$ |
| :--- | :--- |
| C2 $0.02 \mu \mathrm{~F}$ | C10 $0.1 \mu \mathrm{~F}$ |
| C3 $200 \mathrm{pF} 5 \%$ | C11 $200 \mu \mathrm{~F} 12 \mathrm{~V}$ |
| C4 $2 \mu \mathrm{~F} 6 \mathrm{~V}$ | C12 $200 \mu \mathrm{~F} 12 \mathrm{~V}$ |
| C5 $0.01 \mu \mathrm{~F}$ | C13 $320 \mu \mathrm{~F} 6 \mathrm{~V}$ |
| C6 $0.1 \mu \mathrm{~F}$ | C14 $150 \mathrm{pF} 5 \%$ |
| C7 $0.047 \mu \mathrm{~F}$ | C15 30 pF SM |

C8 $0.1 \mu \mathrm{~F}$
C15 30pF SM
VC1-2 208+176pF gang with slow motion and trimmers (TC1-2) (Jackson OO)
TC3-4-5 60pF compression trimmers.
Semiconductors

| Tr1 | BF194 | Tr4 | BC149 | D1 | AA119 |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Tr2 | BF195 | Tr5 | AC127 |  |  |
| Tr3 | BF195 | Tr6 | AC128 |  |  |

## Inductors

Ferrite rod aerial with m.w. and l.w. windings (Denco MW/LW/5FR)
L3 Oscillator (TOC1 Denco)
IFT1-2 (IFT13 Denco) IFT3 (IFT14 Denco)

## Miscellaneous

Speaker $2 \frac{1}{2}$ in. dia. 15 to $30 \Omega$. S2, 4 pole 3 way wafer switch. Veroboard $5 \times 3 \mathrm{in} .0 .15 \mathrm{in}$. matrix. Universal chassis members $6 \times 3 \mathrm{in}$. ( 2 off), $5 \times 2 \mathrm{in}$. (1) (Home Radio). Knobs, materials for case.
extra washers, to avoid shorting or damaging the front section. Two bolts through the circuit board and remaining flange hold these together, soldering tags being fitted for chassis connections, as in Figs. 2 and 3.

If the $0 \cdot 15 \mathrm{in}$. matrix board has 19 rows of holes one way and 32 rows the other, this will allow parts to be placed exactly as shown. The board is $5 \times 3 \mathrm{in}$.


View of completed chassis incorporating band switching.

Prepare it in advance by drilling holes for the bolts and for the pins and screening can tags of L3 and the i.f.t's. A very small round file can be very useful here if any holes are not quite correctly placed. Note that L3 has closer spacing between pins 1-2 and 5-6, and must be fitted as in Fig. 3.

A bare wire, such as 22 s.w.g. tinned copper, is soldered between the tags, as in Fig. 3, for earth returns. One screening can tag of L3 and each i.f.t. is connected to this.

Resistors and capacitors are inserted into the board as in Fig. 2, with their wire ends projecting. The wires are then bent over and connected as in Fig. 3, excess being snipped off. Insulated sleeving should be put on all leads which may touch each other or any metal part. In some places connecting leads have to be soldered on to reach to other components.

It is quite helpful to use black sleeving for the negative leads, red for positive wires and some other colour for other connections, as this makes it easier to check, if necessary.

Leads for the speaker and battery positive can be soldered to pins in these holes, or the wires may project a little and be used as soldering points.

## TRANSISTORS

If $\operatorname{Tr} 1, \operatorname{Tr} 2, \operatorname{Tr} 3$ and $\operatorname{Tr} 4$ are placed as in Fig. 2, the pins will come as in Fig. 3. Leads to these pins should be soldered carefully so as to obtain proper joints, yet avoiding lengthy heating. Note that the BC149 pin connections differ from the BF194/195 types.

Place $\operatorname{Tr} 5$ and $\operatorname{Tr} 6$ so that their lead-out wires emerge as in Fig. 3. These transistors can be left with ${ }^{1}{ }_{2}$ in. or so of lead above the circuit board. Fit diode Dl with the polarity correct as shown.

## FERRITE AERIAL

Two strips of paxolin or other insulating material are cut as in Fig. 2 to raise the rod about $1^{1}{ }_{2}$ in. from the circuit board. These strips are fixed with brackets, and the rod is held with string, thread or elastic through holes and round the rod.

For m.w., the m.w. section only is required, and is placed with the coupling winding L2 away from the rod end, Fig. 2. A and B are the ends of the long winding L1, and C and D are the ends of the smaller winding L2. Thin sleeving is put on the wires and enough slack left to allow the windings to be slid along the rod.

## IF ALIGNMENT

The speaker should be connected to the points as shown. Core settings of the i.f.t's as supplied should be found to be approximately correct, but small adjustments are made for the best results, using a Denco TT5 tool or similar means of rotating the cores.

If a signal generator is available, tune it to 470 kHz and couple it to the base of Trl through a small isolating capacitor, or place an insulated lead near C1. The i.f.t. cores are then adjusted for best volume. If this is done by ear, set VRl at about maximum volume, and keep the generator signal input down so that the actual volume is quite low.


Rear view of receiver. Cut-out in veroboard to clear particular speaker used.

When no signal generator is available, tune in properly any stable transmission and adjust the cores for best results. Repeat with a weak signal.

Once the i.f.t. cores are adjusted for best results, they do not need any further attention and should be left.

## MW ALIGNMENT

If a signal generator is used, band coverage can be set by adjusting TC2 at the h.f. end of the band and the core of L3 at the l.f. end of the band. A small adjustment of L3 core has considerable influence on band coverage.

A signal is then tuned in with $\mathrm{VCl} / 2$ nearly open and TC1 adjusted for best reception. VC1/2 is then adjusted to tune in a signal near the l.f. end of the band and the m.w. winding on the rod moved along to give best results. As these adjustments influence each other they are repeated until no further improvement can be obtained. If alignment is correct adjustment of TCl or the winding on the rod should not give any important increase in sensitivity anywhere throughout the tuning range.


Fig. 4 : Modification of Fig. 1 to permit reception on long waves plus bandspreading of h.f. end of medium wave band.

## LW AND BANDSPREAD

Long waves and bandspreading are added by changing the aerial and oscillator circuit to that shown in Fig. 4. The 4-pole 3-way switch S2 has three positions BS, MW and LW.

In the BS position, S2d places C 15 in series with C3, reducing the coverage obtained with the full swing of VC2. S2a leaves TC5 in circuit, to obtain a suitably reduced tuning coverage in the aerial

For m.w. reception, C15 and TC5 are out of circuit and S2b shorts the l.w. section of the aerial, giving m.w tuning as for the circuit in Fig. 1.

When l.w. reception is wanted, S2b places the m.w. and l.w. aerial windings in series, while S2c brings in the additional capacitor Cl 4 , with parallel trimmer TC3, to shift the oscillator coil coverage.

The tuning ranges obtained were as follows:

$$
\begin{aligned}
& \text { BS }-1800-1425 \mathrm{kHz} . \\
& \text { MW- } 1525-560 \mathrm{kHz} . \\
& \text { LW-280-170kHz. }
\end{aligned}
$$

Any departure from this is not important, however, provided the circuits are adjusted for best results, as described.


Fig. 5: Wiring of the components around the bandswitch shown in Fig. 4.

Fig. 5 shows the wiring of the bandswitch. TC5 is soldered directly from VCl to the switch tag, while TC3 and TC4 occupy the space left on the circuit board.

Place the l.w. section of the aerial winding so that the end having two leads is towards the end of the rod. The twisted connection, consisting of two wires, is tapping $F$ and $D$ of the m.w. section is soldered to this. The single wire at the rod end, point $G$, is soldered to tag MC on VCl/2, Fig. 2. Ends B and E are soldered together and to the tags of $S 2 b$.

Note that the tapping $F$ is electrically near point $G$. There are many more turns from $E$ to $F$, than from $F$ to $G$. The direction in which the turns are wound from A, through to B, to E and through to G, is the same. This will be so if the windings are connected as described. If one winding is reversed, l.w. alignment is impossible. Turns $C$ to $D$ and $F$ to $G$ should also be in the same direction throughout. If necessary leads $C$ and $D$ are reversed to secure
this.


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| KIT |  | tion switch. Mono model E1-23. Stereo $\$ 2 \cdot 24$. Type 2 for ceramic or crystal plekups. Mono 48p. Stereo

gep.

## SWANLEY ELECTRONICS

32 Goldsel Rd., Swanley, Kent.
Mall order only. Postage 37p per order on orders contalning one or more ltems with prlces marked with an". Postage on other ordera 10p.

## 3-BAND ALIGNMENT

With the switch at MW align as previously described. A check can be made that coverage is about $560-1525 \mathrm{kHz}$. If very much out, adjust L3 core and TC2 to correct this. Adjust TC1 at about 1450 kHz and L1 on the rod, at about 625 kHz for best results.
With the switch at $\mathrm{BS}, \mathrm{Cl} 5$ is fixed, so no change of coverage can be arranged, except by altering coverage on m.w. Adjust TC5 at about 1500 kHz for best reception. This adjustment should be found to remain substantially correct through this range but TC5 may be peaked on Luxembourg if wished.
With the switch at LW, adjust TC3 so that the BBC on 200 kHz is about the middle of the band. TC4 may then be peaked at about 250 kHz and the long wave aerial winding slid along the rod, at about 180 kHz , for best results. The position of the core of L3 greatly influences l.w. coverage, so it is necessary either to set this for a m.w. coverage about as mentioned, or to adjust L3 until l.w. reception is suitable, afterwards moving L1 slightly to compensate so that m.w. reception is also satisfactory. All the adjustments can be repeated, in the order given, until no further improvement is possible.


Rear view of the completed receiver showing placement of the battery and speaker.

## INSTABILITY

Commercially made receivers may employ individually selected sets of transistors or make provision for changes to some values, to suit individual transistors, in order to avoid instability.
In this circuit, where oscillation may arise due to particularly "good" transistors, a cure can be effected by adding a ${ }^{1} 4 \mathrm{~W}$ resistor between pin 4 of L3, and pin 2 of i.f.t.l. The lowest value which prevents oscillation is used, but it is not critical. A few spare resistors, such as $220 \Omega, 470 \Omega, 680 \Omega$ or $1 \mathrm{k} \Omega$ can easily be tried. Should the trouble seem difficult to cure, a resistor of some 47 or $100 \Omega$ or so may be placed between point C of L2 and C1. Alternatively, a turn or so may be removed from L2, but this is less easy and difficult to replace.

## COMPLETION

The completed receiver, with speaker and battery, can be fitted in any suitable wooden or plastic case. The case illustrated is quite easily made and is approximately 5 in ; high, $6^{3}{ }_{8} \mathrm{in}$. wide, and $3^{3}{ }_{8} \mathrm{in}$. deep.

Top and bottom consist of "universal chassis" flanged members $6 \times 3 \mathrm{in}$. The top is drilled so that the bushes of the bandswitch, VR1 and VCl/2 spindle, can pass through. Washers or spare nuts are put on the bushes of the switch and VR1 and the receiver is then secured in place by further nuts on these bushes, when the case is finished.

The case sides are ${ }^{\frac{3}{8}} \mathrm{in}$. 3 -ply, each $5 \times 3$ in. They are fixed to the top and bottom by 6BA countersunk bolts through the end flanges. Front and back are similar 3 -ply and are $6^{3}{ }_{8} \times 5$ in.
An aperture is cut in the front panel to match the position and diameter of the speaker cone. Adhesive is spread on the case edges and the front fixed by countersunk bolts through the flanges with a few small panel pins driven through into the wooden sides. The back is held with self-tapping screws into the back flanges.

A piece of speaker material about $5^{1}{ }_{2} \times 3^{1}{ }_{2} \mathrm{in}$. is glued across the front. The front was then covered with self-adhesive material, which can be obtained in many colours and styles. Top and sides are covered in the same way. Strips are cemented on to cover the junction between the speaker gauze and the case covering material.

## TELEVISION

## NOVEMBER ISSUE

## LARGE-SCREEN TV

For large-screen public TV displays it is necessary to adopt some system other than the simple direct viewing of a c.r.t. screen. Many systems have been tried and used and some are suitable for colour TV displays. Development work is still in progress on some techniques. A particularly interesting large-screen colour TV system was used at EXPO 70: this employed mechanical scanning and, as the light source, gas lasers. We shall be looking at this and other systems-such as Eidophor techniquesthis month.

## RENOVATING THE RENTALS

Caleb Bradley tackles another colour chassis widely used for first-generation rental purposes, the Bush-Murphy CTV25/CV2510 series.

## COLOUR RECEIVER CIRCUITS

Gordon King resumes his series, investigating vertical shift techniques and line oscillator circuits, both valve and transistor.


ALTHOUGH desirable for reliability in operation, silicon power transistors were, until recently, rather costly devices. They can now be obtained at prices comparable with germanium types and for this particular project the Newmarket type NKT 0033 have been chosen for the output stage, together with other small signal silicon types for the driver and input stages and for a recommended signal preamplifier.
The power amplifier, the circuit shown in Fig. 1, will deliver 8 W r.m.s. into an $8 \Omega$ load for a THD factor of not greater than $0 \cdot 2$ per cent. The input sensitivity can be varied between 250 mV and 1 V depending on the value of Rb which sets the amount of negative feedback. The output pair are controlled against temperature rise by diodes D 1 and D2 although it is most unlikely that any problem in this direction would occur, providing the heatsink for the output pair is of sufficient area.
The signal input is taken via Trl to the initial driver Tr 2 which is in series with the bias pre-set PRI for setting the d.c. condition of the driver and output stages. The drivers ( $\operatorname{Tr} 3$ and Tr 4 ) are a complementary NPN-PNP pair. Overall feedback is controlled

Fig. 1: The circuil of the main amplifier which uses all silicon transistors and gives an output of 8 W r.m.s.



Fig. 2: (a) The heatsink for the output transistors. (b) An edge view showing the mounting of the circuit board. (c) The siting of the two circuit boards.
by the value of Rb and this also determines the required input sensitivity. The table shows the overall performance obtained with the prototype.

The amplifier is stable on loads down to $4 \Omega$ and will operate at reduced power output (approx 6W) into a $15 \Omega$ load. The quiescent (no signal) current is approximately 25 mA for a supply rail of 30 V . Total current at maximum power output ( $8 \Omega$ load) is about 400 mA at which the supply voltage falls to about 26 V to 27 V .

## CONSTRUCTION

The construction and assembly of the prototype

## $\star$ components list - Main Amplifier




A view of the prototype illustrating the arrangement shown in Fig. 2c.
is shown in Fig. 2. The heatsink must not be less than the area specified, even if an alternative layout and construction is adopted. The assembly shown makes a quite compact unit with the two circuit boards backed onto the main heatsink.

The layout and wiring are shown in Fig. 3a but note that the components are as seen from the outer side of the boards and the wiring from the inner side. The power transistors $\operatorname{Tr} 5$ and $\operatorname{Tr} 6$ must of course be completely insulated from the heatsink with mica spacers and fixing bolt insulators as shown in Fig. 3b. Note also that the two diodes D1 and D2 must be wired so that they can be pushed down to make physical contact with the heatsink. The complementary pair drivers $\operatorname{Tr} 3$ and $\operatorname{Tr} 4$ do not require heatsinks.

## POWER SUPPLY

The circuit and layout are given in Figs. 4 and 5 respectively. The construction of the prototype is on the same lines as the power module with the transformer and rectifier etc. mounted on a piece of aluminium $6 \mathrm{in} \times 3^{3_{4}} \mathrm{in}$. The transformer and rectifier specified will only supply sufficient current for one power amplifier module and a signal amplifier. If two power modules are to be used for stereo operation then the power supply must be able to deliver at least 1 A with a maintained voltage (under maximum

(b)

Fig. 3 : (a) The component wiring of the main amplifier.
(b) The mounting of the output transistors.

Fig. 4 (below): The circuit of the power supply.
Fig. 5 (right): The wiring of the power supply.


Fig. 6: Square wave results on the main amplifier.
power load) of $26-27 \mathrm{~V}$. The same rectifier may be used, as this is rated for 1 A at 30 V but a transformer capable of supplying the higher current will be required.

## POWER AMPLIFIER PERFORMANCE

Before connecting the supply voltage, set PR1 to midway travel. Connect a milliamp meter in series with the positive line and reset VRl to obtain a reading of 25 mA . At maximum power, (just before

040695
To T1 secondary C1 mounted on opposite side



Fig. 7: Oscillogram showing symmetrical clipping.
clipping) the total current should not exceed 400 mA . The input sensitivity can be adjusted by the value of $R b$ and typical settings for this are given in Fig. 1. The overall performance should be as given earlier but square-wave tests will further prove the performance. The result of the usual 1 kHz square-wave test on the prototype is shown in Fig. 6 in which the input is the oscillogram $A$ and the output from the amplifier is oscillogram B. Symmetrical clipping should be obtained at just over maximum (sine-


Table 2. Preamplifier Specification

| Frequency Response | Linear: $\pm 1 \mathrm{~dB}, 10 \mathrm{~Hz}$ to 100 kHz <br> Tape Head: CCIR plus treble lift |
| :---: | :---: |
|  | Mag.P.U.: RIAA |
| Tone Controls | Bass $\pm 14 \mathrm{~dB}$ at 50 Hz Treble $\pm 12 \mathrm{~dB}$ at 10 kHz |
| Distortion Factor Signal-to-Noise | Less than 0.2 per cent <br> -60 dB ref. 400 mV output <br> Tape Head: $68 \mathrm{k} \Omega$, $5-10 \mathrm{mV}$ |
| Inputs (for 400 mV out) | Mag. P.U.: $56 \mathrm{k} \Omega, 5 \mathrm{mV}$ <br> Cer. P.U.: $820 \mathrm{k} \Omega, 80 \mathrm{mV}$ <br> Radio: $200 \mathrm{k} \Omega, 80 \mathrm{mV}$ |

The prototype power supply.

wave) power output as shown in Fig. 7 in which $A$ is a 1 kHz sine-wave input and $B$ the clipped output.

The average d.c. potentials that should be obtained with a 20,000 ohms per volt meter at various points in the amplifier are marked on the circuit; all with respect to common negative.

## PREAMPLIFIER

A circuit for a suitable preamplifier is given in Fig. 8 and this caters for magnetic or ceramic pick-up cartridges, tape head and audio. The radio input can be used for tape replay from a unit with built-in preamp. The performance of the preamplifier is shown in Table 2.

The overall gain has been adjusted to provide 400 mV output to match the input requirements of the power amplifier with Rb (as in Fig. 1) a value of $180 \Omega$. The input circuit $\operatorname{Tr} 1 / \operatorname{Tr} 2$ is quite conventional with selective feedback to provide equalisation for tape head and/or magnetic pick-up. The tape head response has been adjusted to comply with the recording characteristic most frequently used on domestic tape recorders; one which has less treble


Fig. 9: Preamplifier response when switched to Tape Head.
lift than the original C.C.I.R. characteristic. The replay response obtained with the preamplifier is shown in Fig. 9 and provides what is effectively treble lift towards 20 kHz but which is adjusted accordingly for tape speeds of 15 and $71_{2}$ or $3^{3}{ }_{4}$ and


Fig. 10: The preamp response on Mag. P.U. showing R.I. A.A. curve.

## * components list - Preamplifier

| RESISTORS |  |  |  |
| :---: | :---: | :---: | :---: |
| R1 | $68 \mathrm{k} \Omega$ | R16 | 22k $\Omega$ |
| R2 | 56k $\Omega$ | R17 | 10k $\Omega$ |
| R3 | 820k $\Omega$ | R18 | $12 \mathrm{k} \Omega$ |
| R4 | $68 \mathrm{k} \Omega$ | R19 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R5 | $180 \mathrm{k} \Omega$ | R20 | $180 \mathrm{k} \Omega$ |
| R6 | $39 \mathrm{k} \Omega$ | R21 | $120 \mathrm{k} \Omega$ |
| R7 | $680 \Omega$ | R22 | $4.7 \mathrm{k} \Omega$ |
| R8 | $330 \mathrm{k} \Omega$ | R23 | $6.8 \mathrm{k} \Omega$ |
| R9 | $220 \mathrm{k} \Omega$ | R24 | $6.8 \mathrm{k} \Omega$ |
| R10 | 15k $\Omega$ | R25 | $12 \mathrm{k} \Omega$ |
| R11 | $1 \mathrm{k} \Omega$ | R26 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R12 | $3 \cdot 3 \mathrm{k} \Omega$ | R27 | 5.6k $\Omega$ |
| R13 | $15 \mathrm{k} \Omega$ | R28 | $1 \cdot 2 \mathrm{k} \Omega$ |
| R14 | $100 \mathrm{k} \Omega$ | R29 | $1 \cdot 5 \mathrm{M} \Omega$ |

R15 $68 \mathrm{k} \Omega$
All resistors $\frac{1}{4} \mathrm{~W}, 5$ per cent types
VR1 $100 \mathrm{k} \Omega \log$. pot. (volume)
VR2 $100 \mathrm{k} \Omega$ lin. pot. (bass)
VR3 $5 k \Omega$ lin. pot. (treble)

## Capacitors

| C1 | $5 \mu \mathrm{~F} 25 \mathrm{~V}$ min. | C9 | $5 \mu$ |
| :---: | :---: | :---: | :---: |
| C2 | $100 \mu \mathrm{~F} 25 \mathrm{~V}$ min. | C10 | $10 \mu \mathrm{~F} 25 \mathrm{~V}$ min. |
| C3 | $10 \mu \mathrm{~F} 25 \mathrm{~V}$ min. | C11 | 2200 pF |
| C4 | 1500pF | C12 | $0.05 \mu \mathrm{~F}$ |
| C5 | 4700pF | C13 | $0.05 \mu \mathrm{~F}$ |
| C6 | 4700pF | C14 | $100 \mu \mathrm{~F} 35 \mathrm{~V}$ min. |
| C7 | $5 \mu \mathrm{~F} 25 \mathrm{~V}$ min. | C15 | $10 \mu \mathrm{~F} 25 \mathrm{~V}$ min. |
| C8 | 2200pF | C16 | $10 \mu \mathrm{~F} 25 \mathrm{~V}$ min: |
| Semiconductors |  |  |  |
| Tr1 | BC109 | Tr3 | BC108 |
| Tr2 | BC109 | Tr4 | BC108 |
| SW1 | 1 Double pole, | way ro | ary switch |

$17_{8}$ i.p.s. The response for magnetic pick-up is to R.I.A.A. characteristic as shown in Fig. 10.

The output from the first stage ( $\operatorname{Tr} 1 / \mathrm{Tr} 2)$ is taken via the volume control to an emitter follower $\operatorname{Tr} 3$ and then via the active (feedback) tone control network to Tr4.

Care should be taken to prevent hum pick-up in the first stages, particularly in view of the high gain when the tape head or magnetic p.u. inputs are in use. The preamplifier can of course be duplicated for stereo use, in which case a balance control could be included between the ouput and the power amplifier. The insertion of a balance control will of course reduce the overall output signal in which case the feedback resistor Rb in the power amplifier should be about $100 \Omega$. mainly used for local broadcasting there are not many special broadcasts in English to be heard, but the few known to the writer include Radio Sweden on 1178 kHz , Radio Portugal on 755 kHz and 1161 kHz , Deutschlandfunk (West Germany) on 1268 kHz , Radio Pyrgos (Greece) on 1349 kHz , Radio Montecarlo 1466 kHz .

Radio Sweden is on the air nightly at 2245 hr GMT in English and it features a special programme "Sweden Calling DXers" every Tuesday. Radio Portugal transmits the English version of the "Voice of the West" at 2245 hrs , its programme for DXers appearing on a Monday. Deutschlandfunk can be heard every evening at 1745 hrs and there is a fortnightly DX broadcast on a Wednesday. In spite of rumours of a closedown, Radio Pyrgos is still on 1349 kHz . It can be heard in English at 0100 hrs on Mondays and its DX broadcast is on the last Friday of the month at midnight. The Vatican Radio's 250 kW outlet on 1529 kHz is received well in this country and there is a short broadcast in English every evening at 2100 hrs . Trans World Radio uses the Radio Montecarlo transmitter on 1466 kHz for religious broadcasts which can be heard at 2100 hrs .
Harold Emblem of Mirfield in Yorkshire reports reception of the new Voice of America transmitter at Kavalla in Greece which is on 791 kHz . The time of reception was 0300 hrs GMT. At the moment this station is only using 150 kW but it will soon be on full power of 2500 kW when it will become the most powerful medium wave broadcaster in Europe. Harold also mentions the Voice of America at Rhodes on 1259 kHz ; Omdurman, Sudan on 960 kHz with the call "Huna Omdurman"; the BBC Central Mediterranean Relay at Point Delimara, Malta on 1511kHz at 0345 hrs with Big Ben and "Huna London" at the start of the Arabic service.

An interesting letter comes from Norman Burton who lives in Revesby N.S.W. in Australia. He is a keen MW DXer and he says "This place is good for MW DX. I've copied KFSO ( 560 kHz , lkW) in San Francisco at midnight, 8500 miles away. At this hour it is $6 \mathrm{a} . \mathrm{m}$. in Frisco as it is across the international dateline." Norman is constructing the PW MW loop antenna which was described in the November 1966 edition of Practical Wireless. Although this issue is now out of print, copies are available in public reference libraries.

Please send logs and information about the medium waves to the author at 132 Segars Lane, Southport PR8 3JG.

## PRACTICAL WIRELESS will be at THE AUDIO FAIR, OLYMPIA

# A. OF THE CREAT FEATURES In ПEKT MONTH'S PW! 





A great, free competition, open to all readers of PW. Ten (yes, TEN!) Heathkit Oscilloscope Kits Type OS-シ̈ MUST be won. All you have to do is list eight of the excellent features in their order of importance and you could be the winner of a fabulous 'scope. Watch out for the entry forms which will appear in both the December and January issues.

The Music Maker will be one of the most popular features of the year; a portable, keyless, electronic organ which can be built for a song! The prototype has given the P.W. staff endless hours of fun ... and it couldn't be easier: we have arranged for the supply of a special printed circuit board, the case and all unusual components at favourable prices.


By employing a double conversion circuit, this 6-band receiver provides good adjacent channel selectivity, utilising a crystal filter at 455 kHz , while the tuneable first i.f. stage, covering 3.5 to 4 MHz , virtually eliminates image interference. The project can be built in easy stages, initially as a receiver for the 80 m band, later adding a product detector, BFO and crystal controlled front end converter.

You'll certainly be able to stir up the atmosphere with our combined loudhailer and siren... it has all the facilities of commercial models with a lot of extras. Operated from dry batteries inside the main case and using a ready built power module, this is not a complex project. No difficult metalwork is an added attraction to this unusual feature.

# TAKE 2자 JULIAN ANDERSON 

THOSE readers of this column who are blessed with offspring will not need reminding that children love noise. Most of the time they will be discouraged but at parties there is no harm in allowing them to let off a little steam. Our project this month makes an excellent, if noisy, children's party game; it is a variation on the time-tested "Hunt the Thimble" game. Instead of a thimble, we use a small "bleeper" which gives off a high-pitched whistle but at very low level so that it can only be heard in quiet surroundings. The idea is to divide the children into two teams; the organiser chooses the hiding place and when this is done the two teams are set loose. They can either concentrate on finding the bleeper or on making so much noise that the other team will not be able to hear it. Teams can of course divide their forces and have "searchers" and "noisemakers"; it doesn't take much imagination to predict which most children will opt for!

The bleeper, if built up as shown, is very robust and should stand up to the rough handling to which it is bound to be subjected. As with all our projects the cost is low, in this case even new components will set you back only about 50 p plus the battery cost, well worth it for a popular party game.

The circuit is shown in Fig. 1. It is a simple multivibrator working at about 2 kHz although the exact frequency will depend on the particular transistors used.

The sound output is taken to a very cheap crystal earpiece which is wired in parallel with R4. The circuit works perfectly well with a $1 \cdot 5 \mathrm{~V}$ single cell battery but space is a bit limited and a mercury cell Type RM401H may be better. The current drain of the circuit using the specified components is only about $600 \mu \mathrm{~A}$-less than a milliamp-and mercury cells will easily give this.

The operation of the multivibrator is not as simple as many people believe but we have described the operation before and no doubt will do so in the future; for this reason we shall not go into the actual theory. All that you need to know is that it is very easy to build and very reliable.

The prototype was built into a 35 mm film can. A hole can be made in the lid and the crystal earpiece fitted through it. All the crystal earpieces that I have seen are of the same type, though there may be others. In the commonest (if not only available) type the plastic part which fits into the ear screws off and if the hole is made just large enough to take this thread, the lid of the can may be trapped when the earpiece is reassembled. This is illustrated in Fig. 2.

There is no room for a circuit board but there are very few components and these can be wired to each other as shown in Fig. 3. When this has been completed and tested the components can be wrapped in insulating tape to prevent shorting to the can. A switch is shown in the circuit but once again room precludes the use of a conventional type and a wander plug and socket can be used or even wires twisted together and untwisted after use.

# No. 42 <br> "HUNT THE BLEEPER" 



Fig. 1: The circuit is a simple a.f. multivibrator feeding a crystal Earphone earplece.


Fig. 2 (left): The components should be wired to each other as shown. Fig. 3 (right): Shows how the components fit into the can.

## components list

| R1 150kn $5 \%$ +W | 10 |
| :---: | :---: |
| R2 $2 \cdot 7 \mathrm{k} \Omega 5 \%$ it W | 1 p |
| R3 $150 \mathrm{k} \Omega 5 \%$ IW | 1p |
| R4 $2 \cdot 7 \mathrm{k} \Omega 5 \%$, ${ }^{\text {d }}$ W | 1p |
| C1 3300pF ceramic or polystyrene | 3 p |
| C2 3300pF ceramic or polystyrene | 3p |
| Tr1 BC109 | 10p |
| Tr2 BC109 | 10p |
| Crystal earpiece | 20p |
| 35 mm film can | - |
|  | 50p |
| Prices are those recently advertised changed. No allowance is made for costs or for postage and packing and | may have m order points |

Contacts to the mercury cell will have to be made by soldering wires directly to it and this operation should be performed with care. I don't know too much about the chemistry of mercury cells but if the soldering operation is not done quickly the cell appears to discharge almost instantly making the case extremely hot and destroying it, so do take care not to heat the cell up too much.


NEW terminology is being introduced by the progressing art of hi-fi and it is the purpose of this supplement to explain the more important terms used by manufacturers in their specifications as well as some of the established ones found in abundance in the hi-fi literature.

The terms and their definitions, along with sometimes revealing comment, are classified under three headings to which they most appropriately relate. These are Amplifiers, Tuners and Ancillary, the third dealing chiefly with those terms applicable to equipment other than amplifiers and tuners. Sometimes, of course, a term, such as distortion, will be equally applicable under the three headings.

## AMPLIFIERS

## AMPLITUDE DISTORTION

This distortion is present when the amplitude of the output signal fails to follow exactly any increase or decrease in the amplitude of the input signal. It results from non-linearity of the transfer function and gives rise to harmonic and intermodulation distortion. No amplifier is completely free from the effect because its transfer function is slightly curved.

The nature of the curvature determines the order of the distortion produced, but negative feedback and other artifices used in hi-fi amplifiers minimise the curvature within the dynamic range and hence keep the distortion at a very low level.

## ATTACK

This term describes the 'responsiveness' of the amplifier to signals with a fast rise-time, such as produced by percussive sounds of a transient nature. See under RiseTime.

> With Hi-Fi now a booming industry the potential purchaser of audio equipment for home entertainment is faced with a mountain of shiny brochures often describing the merits and specifications of equipment in a jargon he finds difficult to understand. This supplement, compiled by Gordon King, the well-known writer on all types of electronic equipment, is aimed at clarifying the more commonly used terms in the field of $\mathrm{Hi}-\mathrm{Fi}$.

## CHANNEL SEPARATION

This refers to the 'leakage' of signal from one channel of a stereo amplifier into the other channel. The separation is greatest when the signal leakage is the smallest. The leakage signal is commonly expressed as a decibel ratio referred to the full output of the amplifier when the channel in which the leakage signal is measured is correctly terminated both at the input and output. Thus a separation of 40 dB means that the voltage of the leakage signal at the output of the channel tested is 100 times below that of the voltage of the signal at the output of the talking channel.
Owing to capacitive effects, the channel separation sometimes falls with increasing frequency, while impairment at the l.f. end can indicate a common impedance coupling between the two channels. One cause of poor separation is the common impedance presented by the wiper of the balance control on the track of the potentiometer.

## CLASS A

This implies biasing the valves or transistors to the middle parts of their transfer characteristics so that the device is driven upwards on one half cycle and downwards on the other hall cycle. In a push-pull power amplifier, one of the device pair is driven upwards on negative half cycles while its partner is driven downwards, the mode reversing on the positive half cycles. The stage thus draws a constant current at all drive levels within the dynamic range of the amplifier. A class $A$ power amplifier has an efficiency of almost 50 per cent.

## CLASS AB

One type of power amplifier of this class is engineered so that at low drive level the stage operates class $A$, while at increasing drive level the mode changes to class $B$.


The Mullard Pi-Mode class AB power ampllfier design is arranged so that the actual blasing changes from class $A$ to class $B$ with increasing drive.

## CLASS B

The majority of transistor ampliflers adopt this class of power amplifier. The blasing is adjusted such that the push-pull transistors (could be valves) operate at a low no-drive current (called quiescent current). When drive is applied the current in one of the pair rises while the partner is pushed into cut-off on one half cycle, the mode reversing on the other half cycie. Maximum efficiency of a class B power amplifier is 78.5 per cent. Current increases with drive.

## CLIPPING

This is the term used to express the ciipping of the peaks of a waveform when an amplifier is driven beyond its power capacity.


Clipping. The flattening of the tips of a sine wave due to clipping.

## COMPLEMENTARY PUSH-PULL

A power amplifier where the output transistors are of complementary polarities (i.e., p-n-p and n-p-n). In some amplifiers of this kind the driver transistors also constitute a complementary pair.

## CROSSOVER DISTORTION

The type of distortion resulting from class B push-pull power amplifiers owing to the lack of coincidence of the two transfer characteristics at the crossover point. The effect is reduced by applying a critical value of biasing to optimise the quiescent current (see under Class B and Quiescent Current) and hence 'linearise' the middle portion of the transfer characteristic. The situation is further
improved by heavy negative feedback and by circuit design, such that the crossover distortion from hl -fl ampliflers is very small.

## DAMPING FACTOR

This is the ratio of load or speaker impedance to the amplifier's output impedance. Thus the smailer the output impedance the greater the damping factor. The damping factor increases with increase of voltage negative feedback, and with the large amounts of feedback applled to transistor hi-fl ampliflers the source or output Impedance can be as low as 0.1 ohm, giving a damping factor of 80 referred to an 8 -ohm load.
Such a small value of output Impedance Is 'seen' by the speaker as a virtual short-clrcult, excluding the resistance of the speaker connecting cable, and this has the effect of damping the cone movement electromagnetically, thereby minimising the speaker output at the resonance frequencies. The bass resonance is important from this respect, so the damping factor should remain desirably high down to 40 Hz or less.

## DERIVED CENTRE CHANNEL

Some ampliflers are now appearing with a centre channel output derived from the addition of the left and right signals by a matrix. Thls output can be fed to a single-channel power amplifier operating a speaker for location between the left and right stereo speakers.

## DIRECT COUPLING

This refers to interstage couplings or the speaker coupling being direct with no intervening transformer or capacitor. To the speaker it ensures that the damplng factor remains high at l.f. (but Increasing power supply Impedance at l.f. can Influence this), while direct coupling generally minimises l.f. phase shift and encourages enhanced bass performance.

## DYNAMIC RANGE

This is the range of reproduction limited on the one hand by the intrinsic noise of the amplifler (see under Noise) and the amblent background noise level of the listening environment and on the other by the power capacity of the amplifier and speaker system. In many cases the dynamic range is limited by the programme signal, but well recorded gramophone records are capable of a dynamic range approaching 60 dB , which is $106: 1$ power ratio (one million).

## EQUALISATION

When emphasis is given to a range of frequencles during processing or recording (or transmitting), the correction required to normalise the signals during reproduction is called equalisation. Typical is the RIAA recording characteristic which, apart from ensurlng that the recording amplitude is reduced from about 800 Hz downwards, boosts the recording level with increasing frequency so that the required de-emphasis (equalisation) during reproduction not only 'flattens' the overall response but also reduces the record noise at the same time.

## FEEDBACK

This refers to the feedback of the output signal (or a portion of it) to the input. When the feedback signal is negative it subtracts from the input signal and thus reduces the effective output and when it is positive it adds to the input and increases the effective output. Positive feedback produces sustained oscillation or howl. One example is acoustic feedback where sound from the speaker causes


Crossover Distortion. Artifacts on the distortion residual produced by very mild crossover distortion.
response at a gramophone pickup or microphone, a feedback loop then being completed through the amplifier. Howl or oscillation develops at the frequency corresponding to a major resonance in the loop.

## FREQUENCY DISTORTION

This is the unequal amplification of all frequencies over the passband of the amplifier (also see under Frequency Response).

## FREQUENCY RESPONSE

The range of frequencies over which an amplifier responds within defined limits of amplification (or signal output). A parameter expressed as 20 Hz to 20 kHz is the frequency range and is of limited value, while the parameter expressed as 20 Hz to $20 \mathrm{kHz} \pm 1 \mathrm{~dB}$ is the frequency response which means that the deviation of amplification from 20 Hz to 20 kHz is a mere +1 dB or -1 dB , which is much more meaningful.

## HARMONIC DISTORTION

The voltages of harmonics resulting from amplitude distortion expressed as a percentage of the voltage of the fundamental. A common measurement is total harmonic distortion (THD) where the fundamental of a very low distortion sinewave test signal is removed by a steep notch filter, the summed harmonics remaining then being measured as a voltage and expressed as a percentage (or dB value) of the voltage of the fundamental at the required test power of the amplifier.

When measured in this manner the term THD is not really correct since the distortion also has the amplifier noise added to it within the test passband. The correct term is distortion factor.

## HEAT SINK

A metal 'radiator' used in transistor amplifiers to conduct the heat away from the power transistors, thereby allowing them to be used for the delivery of greater power.

## HIGH-PASS FILTER

A filter which, above a critical frequency, allows the unrestricted passage of high-frequency signals. Reciprocally, a bass-cut filter.

## HUM

Spurious signal at the mains frequency ( 50 Hz in the UK) and its harmonics. It is sometimes expressed with noise (see under Noise), so that the total effect is hum and noise.

## HUM-LOOP

A condition arising from the connection of two or more 'earths' to an amplifier system whereby circulating currents of low value at mains frequency and harmonics are added to the programme signals, causing hum to appear in the background.

## INPUT IMPEDANCE

The effective impedance at the input circuits of an amplifier across which the signal source is connected. In this respect the source is loaded by the input impedance. A load sensitive source is a magnetic pickup, most of which are designed for optinum treble response connected to a load of about $47 \mathrm{k} \Omega$. A higher resistance will cause a rise in treble, and a lower one a fall in treble.

## INTEGRATED AMPLIFIER

An amplifier which embodies in a common housing the preamplifier and control section and the power amplifier. Some early amplifiers of large power were in two separate units, one the control unit with preamplifiers and the other the power amplifier.

## INTERMODULATION DISTORTION

Resulting from amplitude distortion (i.e., non-linearity) two or more signals passing through an amplifier intermodulate and hence produce spurious signals, known as intermodulation distortion. Since the distortion components are inharmonious they are subjectively disturbing, more so than low order, even harmonic distortion. Discontinuity in the crossover function of class B amplifiers is responsible for intermodulation distortion and, sometimes, the listening fatigue resulting therefrom. Measurement in based on the voltage of the intermodulation components expressed as a percentage of the voltage of one of the test signals, when two signals are applied to the amplifier in a 4:1 amplitude relationship at a low $(100 \mathrm{~Hz})$ and high ( 5 kHz ) frequency.

## LINE INPUT

Input channel of an amplifier designed to accept signal at a given level from a line at a specific impedance, usually $600 \Omega$.

## LINE OUTPUT

Output channel of an amplifier designed to deliver signal at a given level to a line at a specific impedance, usually 600, .

## LOAD

The circuit or transducer (e.g., speaker) connected to the output of an amplifier. The source (e.g., pickup) is loaded by the amplifier's input impedance (see under Input Inipedance).

## LOUDNESS CONTROL

A control, often the volume control suitably switched, which boosts the bass and to a lesser degree the treble as it is retarded as a means artificially of compensating for the ear's loss of sensitivity at bass and treble frequencies at reducing loudness. A volume control which is permanently arranged to operate in this manner (i.e., loudness effect not switchable) is deprecated by some people and hi-fi enthusiasts.

## LOW-PASS FILTER

A filter which, below a critical frequency, allows unrestricted passage of low-frequency signals. Reciprocally, a treble-cut filter.


## MATCHING

This commonly refers to the selection of the correct impedance to match the source or load and of the correct sensitivity of the input to match the source signal. If the source yields a signal substantially greater than the input sensitivity of the amplifier the volume control will have to be operated at a low setting and overloading of the input preamplifier may result.

## MISMATCH

A condition whereby the coupled impedances inhibit optimum power transference or the source or output signal fails to match the amplifier sensitivity or the load's (i.e., speaker) power capacity.

## MUSIC POWER

Sometimes called dynamic power, and refers to the power that the amplifier (with both channels working, when stereo) is capable of delivering on music typesignal. Output is rated in watts into a specified load value and the test signal is sometimes interrupted sinewave signal (i.e., tone bursts). This method of power expression fails to take account of the power supply impedance and regulation and the efficiency of the power transistor heat sinks. The power obtained is always greater than that measured when the test signal is continuous wave (i.e., steady-state sinewave signal) and with both channels driven together (see under Output Power).

## NOISE

Spurious signal resulting from the random electrical charges existing in conductors, components, valves and transistors. It is heard from the loudspeaker as a 'hiss'. Hum (see under Hum) is also sometimes included as general noise, but since the hum is sometimes filtered in certain measurements, hum and noise may be expressed separately.

## OUTPUT IMPEDANCE

Effective impedance across the output terminals of an amplifier as reflected across the load (i.e., speaker). This impedance is generally much lower than that of the speaker (see under Damping Factor).

## OUTPUT POWER

The maximum amount of power, limited by clipping or a specified value of distortion, the amplifier is capable of delivering to a load of given value (also see under Music Power).
Many transistor amplifiers yield increasing power with reducing load value, and the power is generally given at 4 and $8 \Omega$ (sometimes also at $16 \Omega$ ). The power of stereo amplifiers is given per-channel with one channel only driven and/or with both channels driven together. The almost double demands on the power supply when both channels are driven together reduces the per-channel
power significantly unless the supply is regulated or of a very low impedance when the power amplifiers are class $B$ (see under Class B). However, this does not follow with class $A$ amplifiers (see under Class $A$ ) because the power input remains constant within the dynamic range (see under Dynamic Range) of the amplifier.

The most meaningful power rating is that based on continuous wave (i.e., uninterrupted sinewave) signal with both channels driven to capacity together, referred to a specified level of distortion ( 0.5 per cent or 1 per cent), the power then being calculated by squaring the r.m.s. voltage across the load and dividing by the load value on ohms. This gives the average power per-channel, both running, not the r.m.s. power as often thought and, indeed, specified by many manufacturers!

The peak power on a continuous wave signal is equal to the square of the peak value of the sinewave divided by the load in ohms. Thus when the sinewave passes through zero the power is zero. Hence the average power is equal to half the peak power. The peak power is also equal to the square of the r.m.s. voltage across the load times the square of root-two divided by the load in ohms. Hence half of this, which is then equal to the square of the r.m.s. voltage divided by the load in ohms, is the average power. The power in r.m.s. watts can be obtained by calculation, and it works out to $\mathbf{1} \mathbf{2 2 5}$ times the average power! This is almost 23 per cent higher than the average power. We often wonder whether the spec. writers are aware of this.

## OVERLOAD

When an amplifier is driven beyond its capacity the waveform is clipped and overloading and hence heavy distortion results.

## OVERLOAD MARGIN

It is desirable to have a margin prior to the onset of overload to avoid clipping on transients. This also enhances the reproduction, giving it a 'smoother' quality in many cases.

## POWER BANDWIDTH

The bandwidth over which the full power does not drop below a specified value. The l.f. and h.f. points where the power is half that at 1 kHz (i.e., full power) are often given. This is the half-power bandwidth. Expression is commonly such as ' 20 Hz to $35 \mathrm{kHz}-3 \mathrm{~dB}$ ', the -3 dB indicating a power drop to half the full value.

Sometimes the bandwidth is referred to the low and high terminal frequencies where the distortion fails to rise above a specified value, the DIN value being 1 per cent.

## QUIESCENT CURRENT

The static current drawn by the valves or transistors of a push-pull class $B$ output stage as determined by the biasing in the absence of drive signal. See also under Class B.

## RADIO BREAKTHROUGH

The breakthrough of modulated radio signals into the channels of an audio amplifier owing to the presence of high-level radio signal fields. The effect is that the base/ emitter junction of the low-level input transistor rectifies the signals picked up by the wiring or circuit components and the resulting audio is then handled by the amplifier in the ordinary way so that the radio programmes appear as a disconcerting background on the wanted source signal.

Since the pickup preamplifier has the lowest level input, this source is most troublesome, particularly when the volume control is well advanced. In many cases the trouble can be cured without affecting the amplifier's
response or unduly changing the input impedance (when the preamplifier is in an equalising feedback loop) by connecting a 2000 pF capacitor between the base and emitter of the pickup input transistor, using the shortest possible connecting wires. Both channels of a stereo amplifier must be processed.

## RIAA

Standing for Record Industry Association of America and refers to the recording characteristic of gramophone records. The pickup equalising is a reciprocal of this characteristic.

## RISE-TIME

Time taken by an amplifier for its output to rise from 10 per cent to 90 per cent of its final value when a perfect step (i.e., one with zero rise-time) is applied to the input. The rise-time is a function of the bandwidth.

The upper frequency response of most hi-fi amplifiers, with the treble filter off, is generally sufficient to give a rise-time well in advance of that of mast programme material. (see also under Attack and Transient Distortion).

## ROLL-OFF

The rate of attenuation of a filter at the bass end (highpass) or treble end (low-pass). The crossover frequency is that frequency where the response is -3 dB of the mid-frequency response. The roll-off of an amplifier is similarly defined but in terms of decrease in amplification.

## RUMBLE

Infrasonic disturbance due to the electro-mechanical action of a record player.

## SENSITIVITY

Input signal voltage (r.m.s.) at 1 kHz required to produce maximum amplifier output when the volume control is at maximum and the tone controls and filters set for a 'flat' response and, in the case of stereo amplifiers, with the balance control at neutral.

## SIGNAL/NOISE RATIO

Voltage or power output due to unwanted noise referred to the full power of an amplifier or to some other datum. A $S / \mathrm{N}$ ratio of 60 dB implies that the unwanted noise is $10^{6}$ below the full power or $10^{3}$ below the full voltage of the amplifier. The noise may or may not include hum. This is sometimes made clear by the parameter being expressed such as 'hum and noise 60 dB below full output'.

## TRANSIENT

Signal component of fast rising leading side and usually of short duration.

## TRANSIENT DISTORTION

Distortion of a transient signal component such as produced by resonance and lack of damping, the effect being overshoot or ringing following the fast rising leading side of the signal component.

## WATTS RMS

A commonly incorrectly used term for continuous wave power (see under Output Power).

## WEIGHTED

Often applies to the filtering of a disturbance signal (such as noise) to take account of its annoyance value rather than its absolute value. A weighted $S / N$ ratio is
always better than that obtained from the same amplifier (etc.) when unweighted.

## TUNERS

## AMPLITUDE MODULATION (AM)

Where the audio (or other) information applied to a radio wave causes changes in amplitude. Used generally in the long, medium and short wavebands but is not a hi-fi system as currently employed.

## ATTENUATOR

Network for reducing signal level. Sometimes necessary in the aerial circuit of a tuner to avoid overloading by strong, local signals.

## AUTOMATIC FREQUENCY CORRECTION (AFC)

Often termed automatic frequency control and is a circuit arrangement used in f.m. tuners to correct automatically for any initial, long or medium term tuning error. Currently based on a capacitor diode in the oscillator tuning whose reverse bias is adjusted by a potential delivered by the f.m. detector when the tuning is in error. In some applications it is desirable for the correction to be switchable to avoid the tuning pulling on to a strong station when it is required to receive an adjacent weaker one.

## AUTOMATIC GAIN CONTROL (AGC)

Circuit arrangement which yields a gain reducing bias with increase in signal strength as a method of keeping the output of the tuner constant irrespective of the strength of the received signal.

## BALUN

Usually a transformer designed to accept $75 \Omega$ unbalanced input (coaxial cable) and deliver the signal at $300 \Omega$ balanced (twin feeder). Usable in the converse sense, and sometimes necessary for matching a tuner with $300 \Omega$ balanced aerial terminals to $75 \Omega$ coaxial feeder.

## CAPTURE EFFECT

Effect characteristic of the f.m. system whereby the stronger of two signals, even when in the same channel, 'captures' the tuner so that it responds mostly to this one, pushing the weaker unwanted one well into the background.


Transient Distortion. Example of transient distortion where the fast leading edge of a squarewave initiates ringing.


## CAPTURERATIO

IHF term which refers to the power ratio of two signals in the same channel required to keep the signal/interference ratio to a value of 30 dB referred to 100 per cent modulation and 1 mV input signal level. The ratio of the powers of the two input signals is expressed in decibels, the smaller the dB number the better the capture ratio. Top-flight tuners have a value as low as 1 dB .

## DECODER

Device used in stereo tuners for yielding the left and right channel signals from the multiplex signal fed in from the f.m. detector.

## DISTORTION

Harmonic distortion produced by tuners is generally given at 100 per cent or 30 per cent modulation and 1 kHz . F.M. detector non-linearity may cause the distortion to rise with increasing modulation percentage. Owing to the attenuation of the higher-order harmonics, the THD may beless from tuners with the American $75 \mu \mathrm{~S}$ de-emphasis than from those based on the British (and European) $50 \mu \mathrm{~S}$ deemphasis. Distortion on stereo is also usually higher than on mono.

## FIELD EFFECT TRANSISTOR (FET)

Transistor whereby the current is controlled by electrostatic means. These devices have greater overload immunity than bipolar transistors and a square-law characteristic (minimising third harmonic components) and for these reasons are favoured by many designers for use in the r.f. amplifier and mixer stages.

## FREQUENCY MODULATION (FM)

Where the audio (or other) information applied to a radio wave (i.e., carrier wave) causes changes in the instantaneous frequency of the wave at a rate determined by the modulation frequency and by an amount determined by its magnitude. 100 per cent f.m. corresponds to $\pm 75 \mathrm{kHz}$ deviation on the broadcasting system. F.M. is a $\overline{\mathrm{hi}}$-fi system of broadcasting also allowing the addition (in a subchannel) of stereo information. See under Decoder.

## IMAGE REJECTION

This refers to the rejection by the tuner of a signal at the image (second channel) frequency, corresponding to the tuned (real) frequency plus twice the intermediatefrequency when the local oscillator is working above the signal frequency or minus twice the intermediatefrequency when the local oscillator is working below the signal frequency.

The IHF image rejection ratio refers to the ratio (in decibels) of the signal required for a $30 \mathrm{~dB} \mathrm{~S} / \mathrm{N}$ ratio to that required for the same ratio but at the image frequency. Front-end selectivity increases the ratio.

## INTERMEDIATE FREQUENCY (IF)

The fixed frequency resulting from heterodyning (i.e., 'beating' or modulating to develop the sum or difference frequency signal) the incoming signal with a signal from the local oscillator. The i.f. used in f.m. tuners is commonly 10.7 MHz and in a.m. tuners 470 kHz . The i.f. signal is amplified in the i.f. channel and it is here where most of the selectivity is introduced by tuned bandpass couplings and/or crystal or ceramic filters.

## INTERMODULATION

This can result in the tuner front-end owing to the application of two or more powerful signals. In the group of the three f.m. stations (Radios 2, 3 and 4), third-order intermodulation can put interference on Radio 3. The solution lies in reducing the strength of the signals by attenuation.

## MULTIPATH DISTORTION/RECEPTION

Owing to reflection of the v.h.f. f.m. signal by large buildings, hills, etc., the tuner sometimes receives not only the direct signal but also a reflected signal slightly later due to the greater path distance travelled. This results in high-order harmonic distortion and impairment to the stereo separation and quality.

## MULTIPLEX (MPX)

The name sometimes given to the stereo system of broadcasting and the complex signal delivered by the f.m. detector for application to the stereo decoder.

## PILOT TONE

The signal included with the stereo information, and applied in a subchannel on the f.m. signal, which is used to reform the suppressed subcarrier and synchronise the switching of the detectors in the stereo decoder. Pilot tone frequency is 19 kHz , and doubling this ( 38 kHz ) gives the subcarrier frequency. The pilot tone also activates the stereo indicator light switching.

## QUIETENING (MUTING)

A muting circuit which 'quietens' the audio channel when tuning between f.m. stations and which is 'cancelled' when the incoming signal is of a pre-determined (preset) level.

## QUIETING

The amount (expressed in decibels) by which the noise output of an f.m. tuner is reduced when the input signal is at a given level (see also under $S / N$ Ratio and Sensitivity).

## SELECTIVITY

The ability of a tuner to select the wanted frequency without interference from signals at adjacent frequencies. Front-end selectivity enhances the image (and i.f.) rejection ratio, while good i.f. channel selectivity prevents interference from unwanted stations at frequencies near to the tuned frequency.

Good i.f. selectivity is required by f.m. tuners to minimise 'birdies' interference on stereo reception, coupled with low-pass filtering (critical frequency about 53 kHz ) between the f.m. detector and stereo decoder.

The IHF selectivity, given in many tuner specs., refers to the ratio of the level of a signal in the alternate channel $(400 \mathrm{kHz}$ away from the required station) to the level of the wanted signal for a signal/interference ratio of 30 dB when the wanted signal is applied at a level of $100 \mu \mathrm{~V}$. The signal ratio is given in decibels, and the larger the number the
better the selectivity. Top-flight tuners with ceramic or crystal i.f. filters can have an IHF selectivity better than 50 dB .

## SENSITIVITY

In general refers to the aerial input signal required for a given $\mathrm{S} / \mathrm{N}$ ratio or quieting ratio (see S/N Ratio). The IHF usable sensitivity, frequently given in tuner specs., Is the least aerial signal required across the tuner aerial terminals (p.d.) for a signal/nolse plus distortion ratio of 30 dB . Instead of the modulation being switched off It is notched out by a steep filter so that only the distortion and noise remalns. The test is made with 100 per cent modulation.


Sensitivity. Graphical representation of IHF usable sensit/vity.

## SIGNAL/NOISE (S/N) RATIO

The ratlo of the output signal to the noise at any given level of aerial signal. Signal modulation is either 30 per cent or 100, the latter giving an improvement over the former of about 9dB. A OdB datum is established on the modulation, which is then switched off and the remalning noise measured. See also under Quieting and Sensitivity.


Signal/Noise Ratio. Curve $A$ is tuner output referred to OdB and $100 \%$ modulation. Curve B is nolse output without modulatlon. Max. SiN ratlo about $63 d B$ while input of $1.4 \mu \mathrm{~V}$ results in $S / N$ ratlo of 30 dB .

## SPURIOUS RESPONSE REJECTION

The ability of an f.m. tuner to reject spurious signals falling outside the tuned frequency or the immunity of the tuner itself to the production of spurious signals as the result of intermodulation, etc. (see under Insermodulation).

## SUBCARRIER

The carrier used in stereo broadcasting to accommodate the subchannel stereo components. Frequency is 38 kHz and is suppressed at the transmitter, leaving only the
sidebands, but is reformed at the receiver for detection of the stereo components by doubling the synchronised 19 kHz pllot tone (see under Pilot Tone).

## TUNING INDICATOR

Often a meter which gives both an impression of the signal strength and the correct tuning point at maximum deflection. As maximum $\mathrm{S} / \mathrm{N}$ ratio and least distortion occurs from an f.m. tuner only when the station is accurately tuned, a centre-zero meter might also be incorporated, and the tuning is carefully adjusted until this shows the correct condition of balance.

## ANCILLARY

## AMBIENCE

The term sometimes given to the 'colouration' of the signals due to the concert hall, etc. A kind of 'ambience' from stereo signals can sometimes be obtained by feeding the difference signals (appearing across the two 'live' terminals of the left and right channels) to a speaker sited behind the listener. More advanced matrices are necessary for full effect.

## ANTI.SKATING BIAS

A bias force applied to a pivoted plckup arm to counteract the Inward force (towards the centre of the record) resulting from the drag of the stylus in the groove and the offset angle of the head.

## AUXILIARY BASS RADIATOR

A parastic (non-eiectrically-driven) unit resembling a bass speaker unit located in a loudspeaker enclosure, as If it were an ordinary unit, to increase the movement of alr and hence enhance the bass performance at a given enclosure volume.

## AZIMUTH

Referring to the vertical setting (allgnment) of the head in a tape recorder.

## CCIR

Short for International Radio Consultative Committee, responsible for certain audio standards and tape play characteristics, among other matters.

## CERAMIC PICKUP

A pickup whose generator system is based on plezo electricity produced by the stressing of natural and manpolarised crystals.

## COMPLIANCE

Opposite to stiffness. A mechanical unit whose electrical analog is capacitance. In gramophone pickups one 'compliance unit' is taken as $10^{-6} \mathrm{~cm} /$ dyne.

## CONE BREAKUP

The inability of a speaker cone to work as a piston at high frequencies, the effect being that the cone is not under the complete control of the speech coil, certain parts of it moving in opposition to other parts like a 'rippled rope'. Responsible for uneven frequency response.

## DE-EMPHASIS

A method of signal equalisation where pre-emphasis is used during recording or transmission. Such as the pre-emphasis of the treble modulation on f.m. and the equalising de-emphasis at the tuner. Used essentially as a noise reducing artifice.


## DOLBY SYSTEM

A noise reducing system whereby the recorded signal is 'compressed' and the replay signal 'expanded'. Originally used professionally, but a simplified version for domestic applications (for tape recorders, especially those based on cassettes) has recently been developed.

## DROP-OUT

Short pause in tape replay due to bad tape coating.

## DUAL CONE

Speaker unit containing a main cone for bass and middle frequencies and a smaller, stiffer inner cone for treble frequencies, sometimes called a 'full range' speaker unit.

## DYNE

Unit of force equal to $1 / 980$ gram at latitude $45^{\circ}$ at sea level. Used in pickup mechanics especially.

## ELECTROMOTIVE FORCE (EMF)

Electrical pressure at source. Not to be confused with potential difference (p.d.) which is the voltage developed across a resistance or impedance due to current flowing through it. Both are measured in volts.

## FLUTTER

Cyclic speed variation of tape recorder (e.g. above about 10 Hz ) producing spurious signal related to the frequency.

## HEADSHELL

The end of a pickup arm where the cartridge fits. Sometimes bonded to the arm, though often detachable.

## INERTIA

Resulting from mass and inhibiting change in velocity. Important in pickup mechanics.

## INFINITE BAFFLE SPEAKER SYSTEM

A speaker system where the bass driver is located in an almost airtight enclosure.

## LABYRINTH

A type of speaker enclosure where a maze-like construction extends the air column. Resonances are tamed by heavy damping material.

## MASS

The bulk of matter though not necessarily equal to its weight. A mechanical unit whose electrical analog is inductance.

## OMNIDIRECTIONAL

Having no particular direction of maximum emission or sensitivity. Omnidirectional speaker is one which radiates theoretically equally in all directions.

## PARALLEL TRACKING ARM

Pickup system which allows the cartridge to track on the true radius of the record, as the recording was made, thereby minimising lateral tracking error.

## PINCH EFFECT

Slight variation in the width of the groove resulting from the mechanics of recording which cause the pickup stylus to vibrate vertically at twice the recorded frequency, giving second harmonic distortion.

## PLAYING WEIGHT

The downward force required on the pickup stylus to keep it in the groove and to counter the mechanical reactions of replay.

## PRE-DISTORTION

A scheme sometimes adopted during recording to nullify the effect of tracing distortion on replay. The distortion deliberately introduced is the reciprocal of that produced on replay.

## PRE-EMPHASIS

Deliberate boosting of high frequency to allow attenuation on replay (or reception in the case of f.m.) as a noise reducing artifice. See also under De-Emphasis.

## QUADRAPHONY

A scheme of extended 'stereo' whereby ambient and dimensional information is fed directly or via a matrix to a set of four speaker systems suitably orientated in the listening room. Two main systems are currently under development. One based on four discrete channels and the other on two channels into which are matrixed the information of the other two channels. In the latter system a reciprocal matrix at the reproducing end delivers a version of the original four channels for application to separate amplifiers and speaker systems.

## REFLEX SPEAKER

This differs from an infinite baffle enclosure in that an air vent is incorporated which is tuned by its dimensions for the best bass performance in conjunction with the bass driver and the volume of the enclosure. The auxiliary bass radiator is a development of this technique.

## TIP MASS

The effective mass at the tip of the stylus of a pickup cartridge. Modern techniques have reduced this mass towards 1 milligram or below.

## TRACKABILITY

How well the pickup system will track high amplitude and velocity modulation at a given tracking weight.

## UNIPIVOT PICKUP ARM

An arm which secures both the lateral and vertical movements from a single highly engineered pivot. Difficult to engineer, but characteristics (i.e., very low bearing friction) desirable when well engineered.

## VELOCITY PICKUP

Magnetic pickup whose output increases with increase of recorded velocity. This differs from the ceramic pickup whose output is a function of amplitude or deflection of the stylus.

## wow

The complement of flutter, but at frequencies below about 10 Hz .

# IC SIAMBILISEID POWTR SUPPDIIES <br> <br> D.J.SILVESTER 

 <br> <br> D.J.SILVESTER}

THE design and construction of a compact high stability power supply results in a number of difficulties when the use of a $4^{1}{ }_{2}$ in. $\times 3^{1}{ }_{2}$ in. $x$ 2 in diecast aluminium box is contemplated. This particular supply unit has obvious uses as a 9 V battery replacement, and since output current limiting can easily be added, the supply unit can be used in radio repair work where a current overload would otherwise cause irreparable damage. There is a limitation to the size of the mains transformer due to the smallness of the box, resulting in large variations of unregulated d.c. voltage, a difficulty which can be overcome.
The simple transistor-assisted zener circuit proved to be unstable because of these large voltage variations and the usual multi-transistor feedback circuits needed to give a stable output required more space than was available. The next step was to examine the possibility of using an integrated circuit as the feedback amplifier, and after this was done successfully a second and more powerful supply was built.

## PRINCIPLE OF OPERATION

The block diagram of Fig. 1 shows the basic circuit of the two power supplies. The principle of negative feedback requires that part of the output voltage be compared with a reference voltage, and the output regulator drive adjusted so that the two voltages become equal. The amount of correction applied to the output regulator is controlled by the gain of the


Front view of the completed 9 V supply.
amplifier. It can be shown that the stability of the regulated output depends upon:-

1. The gain of the amplifier.
2. The variations of the amplifier output with changes in the unregulated d.c. supply.
3. The stability of the reference voltage.

The integrated circuit has a very high gain ( 100,000 times) and has been designed so that a change of 1 V in the unregulated supply will cause only a $30 \mu \mathrm{~V}$ change in the output. The first two

fig. 1: Block diagram showing the various feedback paths of the stabilised supply.
sources of design difficulty are eliminated in practice if an i.c. is used as the feedback amplifier. The third source of instability due to the internal resistance of the zener diode can be overcome by providing a stable bias current from the regulated output, and by choosing a diode with a low internal resistance.

## FIXED

 9 volts supplyThe final circuit used for the smaller supply is shown in Fig. 2, the few changes required to give a higher output power will be discussed later. It is possible to use either a $\mu \mathrm{A} 709 \mathrm{C}$ or $\mu \mathrm{A} 741 \mathrm{C}$ integrated circuit as the feedback amplifier, and as five possible package alternatives exist, this component has been represented by the box with connections A to E. These alternatives, as well as the three extra frequency stabilising components required with the $\mu \mathrm{A} 709 \mathrm{C}$ are shown in Fig. 3. It is preferable to use the $\mu \mathrm{A} 741 \mathrm{C}$ integrated circuit since this device will be undamaged should the inputs or


Fig. 2: Above, complete circuit of the fixed 9V supply. Fig. 3: left,
 connections of the various forms of the integrated circuit that can be used. Fig. 4: below, shows the veroboard layout for the smaller components. The 5000 pF capacitors and $1.5 \mathrm{k} \Omega$ resistor (see text) were fitted on the prototype.

## components list



Red and black insulated terminals. Die cast box, 4 x $3 \frac{5}{2} \times 2 \frac{1}{4} \mathrm{ln}$. Mains input plug and socket. Fuse 500 mA and holder. Veroboard $3 \times 1$ 1/in, $0 \cdot 1 \mathrm{in}$. matrix. insulating kit for 2N3055.


010687




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[^4]

Fig. 5: Variation of circuit shown in Fig. 2 to permit the use of p.n.p. transistors such as may be found in an existing unstabilised supply.
output be shorted to either chassis or the unregulated d.c. supply.

The current limiting device consisting of R3 and Tr 2 can be omitted if this facility is not required. The short circuit current with the components chosen is 250 mA . The potentiometer VR1 is used in place of a single resistor in the output sampler, so that the output voltage may be set accurately to 9 V when' construction has been completed. It was not necessary to reduce the sensitivity of this control by using a three resistor potential divider.

## CONSTRUCTION

Very little information is required on the layout of the circuit components, which, with the exception of Trl', R3 and the transformer, can be assembled on a very small piece of veroboard (Fig. 4). One idea which greatly helps the final wiring is that all the components and board are mounted on the lid of the box. A neater appearance results if the larger components and the circuit board are stuck to the lid with epoxy resin adhesive.


Inside the fixed voltage supply unit. The transformer is fixed to the lid with epoxy resin as are short lengths of studding to locate the clrcuit board.

The power transistor uses the box as a heat sink but it must be electrically isolated with washers. The wiring should be thoroughly checked before switching on, after which the output can be set to 9 V by adjusting VR1. A $47 \Omega$ resistor connected across the output should not cause any change in the output voltage.

## VARIATIONS

It is possible to alter an existing power supply having germanium p.n.p. transistors, by using a variation of the original circuit, shown in Fig. 5. The unregulated d.c, supply must not exceed 36 V or the i.c. will be damaged. If a different value of output voltage is required the values of R1, and VR1 and R3 must be recalculated.

If only a small current is required from the regulated output the power transistor can be replaced by a BFY51 or removed completely if an output of less than 10 mA is required.

## VARIABLE voltage supply

After the successful construction of the 9 V fixed supply it was decided that a much more versatile supply could be constructed using the same basic circuitry. This new supply would be capable of powéring small hi-fi amplifiers, and it was decided that in order to increase its usefulness the supply should be able to deliver a constant voltage in the range 10 to 25 V at a maximum current of 1 A . A few small design changes must be made to the circuit of the fixed supply, but one difficulty, the large variations in the unregulated d.c. supply, is to some extent overcome. The final circuit of the second supply is shown in Fig. 6.

The transformer used gave between 20 and 23 V a.c. to the bridge rectifier consisting of four 3 A 50 V diodes, D1 to D4. These diodes produce a d.c. voltage of between 30 and 36 V across the capacitor Cl , the value of this component being increased to $5000 \mu \mathrm{~F}$ by using two separate $2500 \mu \mathrm{~F}$ electrolytics in parallel.


Fig. 6: Circuit of the variable voltage power supply which includes a meter to monitor the actual output voltage.

It should be noted that on no account, or under any load conditions, must the voltage across Cl exceed 36 V or the i.c. will be destroyed.

## OUTPUT CURRENT REGULATOR

The first transistor used for the current output stage is again a 2 N3055 being the cheapest of a number of alternatives. The current gain of this transistor is about 20 , which means that at maximum output current this transistor will require a base current of 50 mA which is greater than the i.c. can supply. The final output current stage consists of a darlington pair Tr1, Tr2 using the 2N3055 and 2N3704/BFY51 transistors. It is calculated that the maximum current demand on the i.c. is now reduced to below 1 mA . The value of the limited output current is $1 \cdot 2 \mathrm{~A}$, obtained by reducing the value of R3 to half an ohm.

## REFERENCE VOLTAGE

The reference voltage is derived from a zener diode D6, and as the stability of this component directly affects the stability of the output voltage
some method of supplying D6 with a constant current must be found. The zener cannot be supplied directly from the regulated output so the most obvious alternative is to provide a stable d.c. voltage from a second zener diode, D5, biased from the unregulated supply. The amplifier chosen is the $\mu$ A 741 C type used in the fixed supply but any of the alternatives given in Fig. 3 may be used if required.

## OUTPUT VOLTAGE SAMPLER

This part of the circuit consists of a resistor chain across the regulated output. The minimum output voltage occurs when the inverting input (lettered A in Fig. 3) is connected to the output rail, the output voltage being equal to the zener voltage of D6. The maximum output voltage is determined by the values of VR1 and R4. The value of R4 is calculated to be $1.3 \mathrm{k} \Omega$ for an output of 30 V . The preferred value nearest to this, $1 \cdot 5 \mathrm{k} \Omega$, produced a drop in the maximum output voltage. There is a rapid increase in voltage with rotation of the potentiometer at the high voltage end, and although this causes some slight inconvenience it was not considered necessary to use separate resistor chains.
-continued on page 642


The smaller components are mounted on a vertical board shown above the large capacitors C1 and C2.

## Only knocks

I would like to comment on the correspondence regarding v.h.f. radio in your October issue.

Mr. James W. Robson's technical explanation for v.h.f. not having caught on is ingenious, but I fear, nonsensical. When f.m. was introduced in this country, and for several years after, only mains driven equipment was available for its reception. In those years, given suitable incentives, a large proportion of listeners could have been persuaded to buy f.m. radios. As it was, shackled by a monopoly broadcasting service, v.h.f. never stood a chance.

In the preface to "Radio \& Television Servicing, 1955-56", a purely technical publication, the editors wrote: ". . . The outstanding feature of this period has undoubtedly been the inclusion, in the ranges of most manufacturers, of combined a.m./f.m. receivers for the reception of the BBC very-high-frequency (Band 2) transmissions . . ." In a subsequent edition it was stated that sales resistance to v.h.f. only sets was strong. and that the absence of Luxembourg on this band was partly to blame. And that is the situation in a nutshell. If only the - opportunity had been taken at the time of the new channels becoming available to introduce independant radio stations, listeners would have clamoured for sets to receive them. Few saw the point of spending money to obtain carbon copies of existing programmes. A parallel may be drawn with the introduction of independent television on Band 3. In the two years following the opening of the Lichfield transmitter I supplied and installed upwards of 1,000 Band 3 converters. In the first two years of the $\mathrm{BBC}-2$ 625 -line service I had orders for just two conversion kits. Even to this day $I$ find customers with facilities for v.h.f. radio and 625 line television who have no intention of using them. What a monu-) ment to wasted opportunity!

One might have thought that the instant and overwhelming success of the so-called "Pirate" radio stations in 1965 would be construed as a clear indication that something was seriouly
wrong with broadcasting in this country. Attempts to legalise the pirates were frustrated by the BBC, who solemnly announced that it was quite impossible to find space for any more stations on the medium wave band. No one appears to have though of offering the largely unused v.h.f. band. Now the BBC has done a complete about turn by foisting its local radio stations onto medium waves. Nothing could be a greater admission of failure on the part of monopoly broadcasting to attract listeners to v.h.f.

The promised commercial stations may inject some life into v.h.f. reception, if only because all will be capable of transmitting in stereo, but by and large I am compelled to admit that I am far from sanguine. Even Lord Hill, contemplating the prospect of nedium-wave reception in the future, is quoted as saying, .. . . and may the Iord have mercy on you after dark!" Unfortunately, f.m. broadcasting missed the boat 17 years ago, and opportunity, as they say, only knocks once. - Charles Miller (Uttoxeter).

## Local radio

No doubt many readers will have noticed the additional interference caused in the m.w. band by the opening of local radio station m.w. transmitters recently, by the BBC. These transmissions, at night, can scarcely be described as local, with a range of over 200 miles. In my location, Radio London is very powerful, and has been received at night in Holland, France and Northern England. This rather defeats the BBC's good original object of a local radio service. Could not the BBC shut down the 206 metres Radio London m.w. outlet, and other outlets, at night, when local radio simply relays programmes carried by Radios One and Two. The v.h.f. outlets could of course remain, as they have done previously. This step would reduce the amount of interference with foreign services, and would bring BBC local radio in line with the proposed restrictions to be imposed upon local commercial radio stations.-John Manners (Middlesex).

## Quadraphonics

I am interested in building my: self a Quadraphonic Hi Fi system.

The design of the preamplifiers and power amplifiers is relatively straightforward, but as yet I have been unable to find any details on how to design the four channel matrix decoder.

An article in your magazine on either the principle of operation or a practical circuit would be of considerable value to myself and I am sure to other people.-- P. M. Cunnington (Essex).

Your letter has been forwarded to me by the Editor and I must comment, with a copy to the Editor, in the hope he may publish, for your letter raises one or two interesting points.

First, before we could contemplate publishing the circuit for a decoder, we must know something of the encoding. Quadraphonics is very much in a state of flux.

In April, there was a show at Osaka, and about half of the 38 exhibitors had some sort of fourchannel exhibition. Few were convincing. Most annoying was the fact that few were mutually com-patible-nor even tried to be!

There are, basically, three systems of stereo recording: the CD-4 of Matsushita and JVCNivico, with American backing from RCA; the SQ, with CBS and Sony throwing in their weight, to which EMI have now added their backing in this country. And last is what might be called a 'Regular Matrix' system, like the Sansui QS, and several related alternatives.

There is no 'universal' decoder, and until the vast commercial battle is won and there is a 'standard' method of encoding, we should play it very close to the chest. I would advise you, Mr. Cunnington, to forgo the pleasure of anything but purely discreet quadraphonics until your way ahead is sufficiently clear. In any case, you can go ahead with preamps and power amps-and, I hope, separated power supplieswith the knowledge that whatever decoder is appropriate to the software available to us (mostly SQ at present, will soon be made available to the constructor.H. W. Hellyer (Bristol).

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THE r.f. signal generator forming the subject of this article was designed for the alignment of superhet receivers as well as for general purpose usage. The cost should be low as the coils are of the Denco plug-in type and do not form a permanent part of the circuit. This has additional advantages in that range switching is not necessary and all r.f. leads and connections can be kept very short; construction is also simplified greatly by using this method. The coverage of the generator depends entirely on the reader. Two coils will cover the medium waves, long waves and the i.f. of 465 kHz but the prototype also worked on the short waves up to at least 10 MHz ; this however is not so vital as the harmonics of the lower ranges will get up to this frequency and way beyond. The circuit used has proved very reliable and gives a good output.

## THE CIRCUIT

The complete circuit is given in Fig. l and may
be considered in two distinct parts, the r.f. oscillator and the a.f. modulator. The r.f. side is very simple and comprises a tuned circuit ( VCl and La ) with base coupling (Lb) with a feedback winding (Lc). Base bias for the transistor is provided through R1 while Cl decouples the supply line with the take off point being C2 from the collector of Tr1. A BC169C is used, mainly because of the lead configuration; this is a plastic encapsulated version of the BC109. The supply for this stage is provided through a dropping resistor R2.
The a.f. section comprises $\operatorname{Tr} 2$ and associated components. This is a Hartley configuration using a transistor output transformer type LT700. This is a completely conventional arrangement, the primary of the transformer is tuned by C4 with C5 providing the feedback coupling and R3 the base bias to start the operation (this resistor may not be necessary as the feedback should be of sufficient level to maintain oscillation but it is as well to include it to make sure).
The output from the a.f. oscillator is at a fixed frequency-in the case of the prototype it is about 800 Hz but the exact frequency will depend on the particular transformer and on component tolerances etc. The frequency can be increased by lowering the value of C4 or omitting it altogether, relying on the self-capacity of the windings to tune it. The frequency can be lowered by increasing the value of C4.
The waveform from the audio oscillator is far from a pure sine-wave but this is of no consequence and serves the desired purpose admirably.

## FUNCTION SWITCH

Switch SW1 is the main function switch providing the on-off facility, r.f., modulated r.f. or a.f. at the output socket. In position $A$ the battery is unconnected and so the circuit is off. In position $B$ the


Fig. 1: The complete circuit of the R.F. Signal Generator. The required range is selected by plugging in the appropriate coil. SW1a and SW1b are ganged to give the required output.
positive supply is connected between D1 and R2. In this mode the diode will block any current to the a.f. oscillator and this will be inoperative. Current, however, is allowed to reach the r.f. stage and SWIb connects the output level control to the output of Trl via the d.c. blocking capacitors.

In position C, VR1 remains connected to the output of the r.f. section but SWla applies current directly to the a.f. oscillator stage. In this mode Dl allows current to reach Trl but the positive supply is now connected via the secondary of the a.f. oscillator transformer. This creates a small, rapid change in the supply voltage in sympathy with the frequency of operation ( 800 Hz in the case of the prototype) and so the output from Trl will also vary, modulating the signal. Normally it is not good practice to modulate an oscillator as this can cause both amplitude and frequency modulation but in practice some slight f.m. can be ignored. This arrangement gives a modulation depth of about 20 per cent.
In position $D$ the positive supply is as for $C$ but SWlb connects to the collector of Tr2 via the d.c. blocking capacitor C3 giving an audio output. If it is important that only pure a.f. is provided, the r.f. section can be made inoperative simply by unplugging the coil.
Capacitor C6 decouples the supply of the a.f. stage, the relatively low value is quite sufficient because of the audio frequency used.

The output of the r.f. section is about 500 mV and that of the a.f. section about 400 mV ; in both cases this is more than adequate for normal usage and both outputs connect via the level control VR1 before connecting to the coax output socket.

## CONSTRUCTION

The coils used are the Denco plug-in types, these fit into a B9A valve-holder. The leads can be kept very short by building the r.f. section directly onto

## components list

Resistors<br>R1 $68 \mathrm{k} \Omega$<br>R3 180k $\Omega$<br>VR1 $10 \mathrm{k} \Omega$ lin. pot.<br>All fixed resistors $\downarrow \mathrm{W}, 5$ per cent tolerance<br>\section*{Capacitors}<br>C1 $0.05 \mu \mathrm{~F}$ ceramic disc<br>C2 3300pF ceramic, polystyrene or silver mica<br>C3 $0.05 \mu \mathrm{~F}$ ceramic disc<br>C4 $0.05 \mu \mathrm{~F}$ ceramic disc<br>C5 $0.05 \mu \mathrm{~F}$ ceramic disc<br>C6 $10 \mu \mathrm{~F} 12 \mathrm{~V}$ electrolytic<br>(All0.05 $\mu \mathrm{F}$ can be $0.047 \mu \mathrm{~F}$ )<br>VC1 365pF tuning capacitor, Jackson Type 00

## Semiconductors

## Tr1 BC169C <br> D1 1 N4001 or similar <br> Tr2 BC109

## Miscellaneous

Coils: Denco transistor range, Blue, 1T, 2T, 3T or to suit; T1: Eagle transistor output transformer type LT700; SW1: 4-way, 2-pole rotary type; B9A valve holder; Coax output socket; Veroboard 0.15 in . matrix $16 \times 13$ holes; PP3 battery and battery clip; Terry clip for holding battery; Aluminium angle; Plastic sheet for pointer; Knobs to suit; Metal chassis $5 \times 4 \times 2$ in.


Fig. 2: The wiring of the r.f. section on the B9A valve base into which the coil is plugged.
the pins of the valveholder as shown in Fig. 2. To neaten the layout some changes have been made to the recommended Denco connections but the electrical result is the same. To bring the negative chassis connection to this part of the circuit a short wire runs to a solder tag fitted underneath one of the valveholder mounting screws.
Pin 4 of the valveholder does not connect to the coil and is only used as a convenient anchoring point for C2.

The layout and construction of the a.f. section is much less critical than that of the r.f. oscillator and can be built in any convenient form. In the prototype these components are mounted on a piece of Veroboard of standard $2^{1}{ }_{2} \mathrm{in}$. width, 13 holes long. The layout of this section is shown in Fig. 3.


Fig. 3: The component layout of the a.f. section on Veroboard.
Holes should be drilled out to take the mounting screws and the lugs of the transformer. There are two strips on this board which are connected to chassis potential, these are linked together through the transformer casing, the lugs being bent inwards and soldered to the adjacent strip. One of these strips also connects to a link wire which runs to a solder tag fitted under one of the component board mounting screws.

There is no crowding of the components and construction should present no problems.

A few breaks are necessary in the conductor strip; one underneath the transformer is absolutely necessary, the other two will only be needed if the solder tag or the transformer lugs touch an adjacent strip.

Four take-off points are required; these are shown in Fig. 3 and Fig. 4, the latter shows the wiring not dealt with in the other drawings. A stiff wire should


Fig. 4: The wiring between the controls and the two main operative sections.
be used from the fixed vanes of VCl to pin 1 of the valveholder.
The rotary switch used is unlikely to look like that in Fig. 4, it is only shown in this way for clarity. The coax output socket used was of the insulated type and so a chassis connection has to be made to it; more conventional types will not need this.

It is very important that a good quality valveholder is used. Initially the author used rather a poor one and this later had to be replaced. The frequent plugging and unplugging of the coils demands a first rate quality component here.
The pointer can be made from any thin rigid clear plastic; the dimensions for this are given in Fig. 5a. A score mark should be made centrally on the underside, this can be inked in to make it clearer.

The component board is fixed to the chassis by


Fig. 5 : (a) The construction of the frequency pointer. (b) The mounting bracket for the a.f. section component board.
means of an aluminium angle bracket as shown in Fig. 5b.


The screening can lid should be fixed on top of the valve base, ensuring that there is an electrical connectlon.

The main case is a $5 \times 4 \times 2{ }^{1}{ }_{2}$ in. chassis-this is a fairly common size and the drilling of the face is shown in Fig. 6. The only part of this which may cause problems is the mounting of VCl ; the dimensions of these holes and their siting are rather critical. It is best to first drill a ${ }_{2}$ in. diameter hole, this will then accommodate the raised section around the ball race. The capacitor itself can then be used as a template for finding the exact locations of the two mounting screw holes. Very short screws must be used to mount this component, if they are too long they will foul the moving vanes of the capacitor.

The valveholder is mounted centrally on the top


Fig. 6: The drilling details of the chassis face.


Rear view of the completed prototype.
of the case; no problems are likely to be associated with this and so dimensions are not given.

Battery mounting is often a problem; it has been solved in this case by using a Terry clip which grips the PP3 battery; this is fitted behind the level control.

To have control over the output level, the coil should be screened; if it is not it will radiate considerable power. The screening can supplied with the coils may be used for this. A $3_{4} \mathrm{in}$. diameter hole should be cut in the lid and this glued over the valve base at the same time ensuring that there is an electrical connection to chassis. A hole should also be drilled in the top of the can to provide access to the coil core adjusting screw.

## CALIBRATION

Articles on r.f. signal generators frequently make a lot of calibration, this however need present no problems if you have available an all-band radio. Radio stations maintain very accurate frequency control and by beating the r.f. of our signal generator with their carriers, we can obtain very accurate calibration.

A blank calibration panel is given in Fig. 7. This can be cut out and stuck onto the face of the case or copied.

It is best to calibrate the long wave band (range 1) first as this will also give us our i.f. frequency. Assuming that all is well with the circuit and wiring, when the range 1 coil is plugged in and the function switch is set to MOD, the output should be tuneable over the long wave band and this will be heard on a radio. A wire should be connected to the inner of the coax socket and placed near the radio, the level

Easily received MW/LW Radio Stations that man be used for calibration. Extra points can be found by interpolation

| Range 1 |  | Range 2 |  |
| :---: | :---: | :---: | :---: |
| Frequency | Station | Frequency | Station |
| $\begin{aligned} & 151 \\ & 180 \\ & 200 \\ & 233 \\ & 245 \end{aligned}$ | Deutschlandfunk <br> Europe No. 1 <br> BBC Radio 2 <br> Luxembourg <br> Denmark | 529 <br> 602 <br> 647 <br> 701 <br> 746 <br> 800 <br> 908 <br> 1007 <br> 1106 <br> 1196 <br> 1295 <br> 1403 <br> 1502 <br> 1594 | Algeria <br> France <br> BBC Radio 3 <br> Andorra <br> Hilversum 1 <br> W. Germany <br> BBC Radio 4 <br> Hilversum 2 <br> AFN Munich <br> VOA Munich <br> BBC European <br> Service <br> France <br> Warsaw (Otten <br> English in late evenings) <br> BBC Radio <br> Leicester |

should be turned down as low as possible to prevent pulling in the receiver.

The receiver should be tuned to the German Deutschlandfunk programme on 151 kHz -this is usually well received throughout Europe. With the vanes of VC1 fully closed and the pointer lined up with the end of the scale the core of the coil should be adjusted till the a.f. note, combined with a beat note, is heard. When this is done, tune VCl to a point where the vanes are almost completely open. If the receiver is a superhet with an i.f. of 465 kHz (or 455 kHz for many imported sets) another note will be heard. Unlike the 151 kHz point this note will be heard from the receiver irrespective of the band or frequency (though not on v.h.f.) the receiver is tuned to. When this point is found it should be


Fig. 7: This basic scale may be cut out and calibrated as described in the text.
carefully marked as this is the i.f. If possible, two positions should be marked, one for 455 kHz and another for 465 kHz .

Range 2 should be adjusted in a slightly different way. If we adjusted the core of the coil to operate on, say, 529 kHz (Algeria), the high frequency end of the medium wave band would be badly cramped as this point would be reached before the vanes of VC1 reached their "extra spread" setting. For best results the core should be adjusted so that at a frequency of about 700 kHz coincides with the vanes half open (pointer vertical). Andorra on 701 kHz can be used for this. The range should then extend from 500 kHz to over 2 MHz , encompassing the complete medium wave band and the $1 \cdot 6 \mathrm{MHz}$ i.f.

Range 3 should be adjusted so 1.6 MHz coincides with the vanes fully closed. This range should then go up to about 6 MHz .

The harmonics of range 3 can be used to give coverage of up to 30 MHz though these are of less use and will be of lower level.

Various low level whistles may be heard on the wrong frequencies-these can be due to harmonics or receiver action but they should always be at a low level and there should never be any real doubt about determining the fundamental, it will be by far the strongest.

A table of useful stations for helping in the calibration are given; for a more complete list. reference books such as the "Guide to Broadcasting Station" or "World Radio TV Handbook" should be consulted.

P.W. January 1971, Feb. to Dec. 1971 (inclusife) and Jan. 1972.-G. A. Hunt Garrich Close, Ings Road, Hull, Yorkshire.
...P.W. Oct., Nov., Dec. 1974. - Alan Ward, 3 Cressbrook Avenue, North Wingfield, Chesterfield, Derbyshire, 5425 NL .
..P.W. March 1968 and June 1968 wlth Items on "Clubman"' recaiver. Als o required nformation on Lafayette HE 40 radio. Including clrcuit diagram. Willing to loan or buy,-B. S. Smith, "Barnston", 58 Blackburn Road, Rishton, Blackburn, Lance. ... Beginner requires any back numbers of P.W. and P.T. partlcularly P.T. June uly 1970, June/July 1966.-L. Cooke, 21 St. Anthonys Road, Bournemouth, Hants. gatured In that eatured
Yorks.
i..P.W. for Jan, and Feb. 1971 contalning detalls of Stereo tape recorder construc PE. Scon
...P.E. Feb. 1968 containing I.C. tape recorder.-D. Hollis, 15 Merlin Close, Coller , Romford, Essex.
Wanted P.W. October 1971 and Aprll 1972. Postage and cost refunded.-K. Foster,
Malmes bury Road, Shirley, Southampton, SOI 5 EY.
...Feb. and March Issues of 1971 containing the P.W. Stereo Tape Recorder.
J. Thomas, 16 Western Road, Urmston, Lancs. M31 3LF.

Aprll and May 1960 copies of P.W. and any information on 19 set Mk, III.-T. Izard, Clumber Drive, Radclifi-on-Trent, Nots.
ouglaston Crescent, Milngavie, Glasgow, Scot and.. W. W. Jan. 1972 and Feb. 1972 issues.-C. W. Lo, 183 Lodge Avenue, Dagenham,
Essex. W. Feb. to June 1967 and Aug. to Dec. 1970. P.E. Volumes 4, 5, 6 and Feb. 1971 Alf in very clean condition for binding. -E. Somerville, 2 Hillcrest Place, Killwinning Ayrshire
...Popular Hi-fi April, June and October 1971.-R. S. Wood, Preston Secondary chool. Monks Daie, Yeovil, Somerset.
wanted P.W. Feb 1972.-P. Ducker, 50 Cheshire Road, Leicester.
aip. W, Oct. April 1968 issue of P W-C Say is, 60 Hardens Mead Chi Menhad.
P W May anse of P.W.-C. Saylis, 60 Mardens Mead, Chippenham, Wilts winning. Ayrshire, KA1360H.

Jan. 1972 issue of P.W.-S. Bouma, Wild Hoefaan 31, The Hague, Netherlands April 1970, May 1970 and June 1970 P.W.-J. Slim, 1 Whitehall Road, Smethwlck Warley, Worcestershire.
W. P.W. Feb. 1972.-G. Burns, 31 Beechwood Drive, Eaton, Congleton, Cheshire W12 2NQ.
S. Jan. 1972 and Feb. 1972.-B. A. Cotton, 15A. Thomas King House, Weliing Si Woventry.
Hastings, New Zealand (Airmail)
P.W. Feb. 1972.-P. Parker 31 Fair View, Handsacre, Rugeley, Staffs, WS154DJ ... April 1966 P.W. including '19' Set Mods. Article.-A. C. Arthur, Pinkworthy Farm. Pyworthy, Holsworthy, Devon, EX22 64Q.


THE art of sonar is not quite as simple as it sounds. In radar, an electromagnetic wave is transmitted in bursts and the energy from these which is reflected by surrounding objects is viewed on a cathode ray tube. Different objects cause different amounts of reflection and thus cause a pattern to be set up.

Sonar does the same sort of thing but with two major differences. First, it uses sound waves instead of radio waves. Secondly, it functions under water the classic application being in ships.

Basically, sonar sounds simple until one considers it in greater depth-so to speak! A moving target generates acoustic energy, a form of noise, which can be detected. Thus no generated pulses would be needed in this passive mode. Perhaps, already I have stimulated anglers into thoughts of being able to actually "see" the big ones or shoals? Incidentally, professional systems do exist which can detect a single cod at 600ft!

In the case of sonar, the speed of data collection is limited mainly by the velocity of the propagation of acoustic energy through the water. In sea water, this is about 1,500 metres/second. Although this may sound (sorry!) fast, compare it with electromagnetic energy which has a velocity of propagation of some $3 \times 10^{8}$ metres/second. Thus radar would take only $1 / 1,000$ th. of a second to receive a reflection from a target nearly 100 miles away. For sonar, it would take something in the order of four seconds for a target about 3,200 yards away-barely two miles.

Like its r.f. counterpart, sonar has problems with noise. The surface of the sea is in a continual state of variation due to the weather. Marine organisms cause noise, so does seabed turbulence.

At u.h.f., microwave energy in the earth's atmosphere travels in a straight line tending to follow the curvature of the terrain. Unhappily, sonar's acoustic. energy is very reluctant to follow any straight line in sea water.

To complicate matters further, water temperature has an affect. In the sea, there is a temperature gradient. One "layer" of water is a little warmer than the next one down which in turn is a different temperature to the one immediately below it. It is not until nearly a mile down that an almost constant temperature is reached.

It is interesting that the velocity of acoustic energy varies dependant upon the temperature, pressure, salinity etc. Thus the acoustic waves are bent or diffracted (like light waves in water and glass) the amount of diffraction depending upon

# practically wireless commentary by ILIIIII 

SOME while ago, Henry had the uncomfortable pleasure working beside a pedantic colleague who corrected every casual remark as if invested with the guardianship of the English language.

A favourite dampener was: in answer to a query, "What does the meter say?" "The meter doesn't. You have to read it!"

Recent forays into the world of instrumentation have convinced this traveller that we shall soon render Old Niggly out of date. We have lately been promised computer type voltmeters that will store a measurement and notify a comparison. A sort of digital poor cousin of the storage 'scope. It is but a small step to the meter that will shout: "Hey! Give over" when overloaded, or a prim: "Three-pointseven millivolts" at the appropriate moment. With, of course, a built-in recorder (opto-magnetic; J mean to invent it) so that Old Niggly cannot dispute the matter, though he is sure to contest the manner of its delivery.

Avid and persistent collectors of Henribilia will by now have gathered that the subject of digital read-out devices for the ordinary radio workshop occupies the upper layer of the author's


The meter that will shout "Hey! Give over!"
(for want of a better word) mind
It is true. I have recently come into proud possession of a digital multimeter.

Quite apart from a phenomenal accuracy, better than the thickness of a conventional pointer on the lowest voltage or highest resistance scale, the Digivom has the advantage of push-button range selection, and a read-out large enough to satisfy the most myopic niggler.

Mind you, it is not all plain sailing. There are one or two disadvantages the brochure omitted to mention. One is the tendency the Digivom has of sending out its own identifying signal-a series of chopped pulses that render its use as a measuring instrument for delicate audio (and some radio) purposes, about as valid as a Bobby Fischer unpaid checkmate.

The darned thing radiates all through the workshop. And if you enclose it in a tin box, leaving only the display tubes glowing through their visor like the Red Knight's menace, it still manages to jigger the f.m. band on the next-door engineer's rush job.

At the moment, I have a Solartron rival which uses diode display devices and these radiate only a mild hash, spesumably there is hope for we humble servicemen yet. Except that this one has a read-out muči smaller and of a style which, lite the numbers my friendly, neighbourhood bank manager scrawls along the bottom of my cheques, defies analysis except by logical devices.

It is fascinating to watch the digivoms trundling down through the voltages when one reads off the h.t. line. Almost like a Cape Kennedy transmission!

Henry was happy with his pushbuttons till Old Niggly sowed the seed of doubt in his mind. "Why not", he said, "do away with the push-button rangè ${ }^{\text { }}$ selection?" After all, if you can read-off an overload indication and produce


Throw one or 'wo in the makers' faces.
a flashing light, and do a similar service for the wrong d.c. polarity, why not build in a trigger circuit fed from a sensing device so that the meter must always switch itself to a range appropriate to the voltage applied to its terminals?

Well, why not? I asked an electronics designer whose name you will have seen if you read the glossier audio journals or works on instrumentation. He said clamly: :"It has been done," and quoted me one of his own patents.

Now Henry doesn't pretend to be a designer or an inventor, and his only claim to innovatory fame is a range of hook-up switchboxes. But, fired by the ambition engendered by necessity-and not a little niggled by chagrin at spending over a hundred quid on a digivom before discovering its drawbacks-he has sworn to throw one or two modifications in the makers' faces before the year is out. Watch this space for details.

The modifications will not, true, confound Old Niggly by making the machine shout its readings to all and sundry. I cannot hope to incorporate storage facilities. But the least that can be done for an instrument that boasts only one current scale, and that the "basic one", is a switched range of external shunts. That should not be beyond Henry's abilities-when he gets five minutes to spare.


## R

8-track stereo cartridge TAPE DECK MODEL RP-1000ST The popular Lear- Jet type recording unit is the heart of the fantastic AP- 1000 ST . Which has full record
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## ${ }^{2}$

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### 227.05 <br> \&

PALACE AM/FM/MPX STEREO TUNER AMPLIFIER SSA-16 This is one of the lowest priced staro tuner amplifiers on the market. It covers the full range of both $A M$ and $F M$ broadcast requencies. And when you'rs switchad to FM. an indicator ughts of whan a storeo signal is received - that's tha time to swatch to 'Stereo'! The SSA- 16 has all the facilities you'd ex pect to find on tunters costing twise as much - separate vol ume, bass, trable. balance and tuning controls. Selector switch for tape. phono. AM, FM, stereo. Jack sacket on front panel for stereo headphones. Frequency range: $F M 88-108 \mathrm{MHz}_{2}$. AM $535-1605 \mathrm{kHz}$ Frequency response: $50-10.000 \mathrm{~Hz}+3 \mathrm{~dB}$ Power output: 4 watts total music power into two 8 ohm spazkers. Size: $16^{\circ}$ wide. $11^{*}$ high, $8^{\text {r }}$ deep.

compact unit measuring only $5 \frac{10}{3}$ " wide. $1 \frac{13}{3^{-7}}$
high and 6 g" deep. It tontans its own majns
power supply, and has a ganged volume contro! and separate treble controls for each channel. Spacification: frequency response $40.17000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$; output 3.5 watts music power per channet into 8 ohms: input. phono, 60 mmV ; signal $\cdot 10$-noise atio better than 45 dB .

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| 2N4289 |  |  | 150 | BD130 |  |  | 65p |
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| to3 Trangletor Covern, es. 7 p |  |  |  |  |  |  |  |

RESISTORS 10\% $\%-5 \%-2 \%$ Code Power Tolerance Range Valuen 1 to 9 (see note below)

|  |  |  | In olma |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 1/BW | 5\% | 4.7-470K | Re9 | 1 | 0.8 0.8 | 0.7 0.7 |
| c | 1/4W | 10\% | 4.7.10M | E18 | 1 | 0.8 | 0.7 |
| 0 | $1 / 2 \mathrm{~W}$ | B\% | 4.7-10M | 8 EP 4 | $1 \cdot 2$ | 1 | 0.9 |
| 0 | 1 W | 5\% | 4.7.10M | $\underline{818}$ | 2. | 3 | 2 nett |
| 10 | 1/2W | 2\% | ${ }_{0}^{10.22 .3 .9}$ | ${ }_{\mathrm{E} 212}$ | 7 | 7 | ${ }_{6} 6$ |
| WW | 1w | $\begin{aligned} & 10 \% \\ & +1 / 20 \Omega \end{aligned}$ | 0.22.3.9 | E12 | 7 | 7 |  |
| WW | 3W | 5\% | $1 \Omega \cdot 10 \mathrm{~K}$ | E18 | 7 | 7 | 8 |
| WW | 7W | $5 \%$ | $1 \Omega \cdot 10 \mathrm{~K}$ | E12 | 9 |  | 8 |
| Codes : $\mathrm{C}=$ carbon filnu high stablility low nolse MO = metal oxide Electrosif TRs uitra low noise WW:wire wound Plessey |  |  |  |  |  |  |  |
| Yalue E12 their Ol an power | enotes ecades their rating, | ;ies: 10,12 24: as El: cades. Pric OT mired | $2,15,18$, <br> pius 11 , <br> en are in <br> values. | 2, 27 16, nes | 39, |  | 82 and 62, 72, ue and t value |

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| FLH281 | (7442) | 1.16 | FLJ191 | (7495) | 870 |
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# THE BROADCAST BANDS Malcolm Connah 

## Frequencies in kHz - Times in GMT

IT IS a generally accepted axiom that good shortwave reception is 50 per cent dependent on the aerial, the other 50 per cent accounts for the receiver and skill of the operator etc. I know many people who have bought expensive receivers and been disappointed with the results simply because they did not have an effective aerial system.

The simplest form of aerial is the long wire, and the longer the wire is the better the reception will be. The strength of the signal is measured in microvolts per metre; the longer the aerial is, the stronger the signal will be because the voltage induced in the aerial will be higher. All receivers have a certain sensitivity which is a measure of the voltage required at the aerial input to give a useful audio output.

A long-wire aerial of 50 feet will usually give adequate reception whilst one of 100 feet is very good. Other factors affect the performance of such an aerial, consequently the higher above ground it is the better it will perform. Care should also be taken to keep the aerial as far away as possible from buildings, trees and any other obstructions which may reduce the signal induced. It is also essential to keep the aerial as remote as possible from any sources of man-made interference.

A better type of aerial is the dipole. This type of aerial is restricted in use because it can only be tuned to a single band. The advantage of a dipole, however, is that being tuned its performance on the band of interest is superior to that of a long-wire.

The dipole aerial consists of a single length of wire which is cut in half, the two halves are connected via a coaxial cable to the two terminals at the rear of the receiver. The overall lengths for the various Broadcast Bands are as follows:

| Frequency Band | Length of aerial |
| :---: | :---: |
| 25 MHz. | $5 \cdot 60$ metres |
| 21 MHz. | $6 \cdot 65$ metres |
| 17 MHz | $8 \cdot 00$ metres |
| 15 MHz | $9 \cdot 30$ metres |
| 11 MHz | $12 \cdot 10$ metres |
| 9 MHz | $14 \cdot 80$ metres |
| 7 MHz | $20 \cdot 00$ metres |
| 6 MHz. | $23 \cdot 40$ metres |

For the benefit of those who have not been converted to the metric system one metre is approximately equal to $3 \cdot 28$ feet (e.g. $20 \cdot 00$ metres $=65 \cdot 6$ feet).

An interesting feature of the dipole is that it is a directional aerial. If the wire is aligned NorthSouth the best reception will be from due East and due West.

It is possible to use a dipole on all bands but without the same sensitivity as the band to which it
is tuned. For example a 7 MHz . dipole will be a sensitive, directional aerial on the band of interest. but on all other bands it will perform as a nondirectional, 66 foot long-wire.

Most listeners have one band in which they are particularly interested, the ideal solution in this case would be to erect a dipole for this band and a longwire for general listening on the other bands.

The first report this month comes from Michael Berry in Dewsbury, Yorkshire. Michael is still using his Eddystone EB35 receiver but has replaced his vertical aerial with a 20 foot long-wire in the loft. The stations logged included:
3960 Baghdad, Iraq in Arabic at 2310.
4800 R. Lara, Venezuela in Spanish at 0240.
4820 Voz Evangelica, Honduras, English at 0340.
4880 R. Universo, Venezuela, Spanish at 0330.
4990 R. Barquisimeto, Venezuela at 0230.
4995 R. Brazil Central in Portuguese at 0110.
5075 R. Sutatenza, Columbia at 0100 .
Richard Keen of Hayes in Middlesex has just erected a long-wire of 120 feet which enabled him to hear:
4790 R. Ondas Portenas, Venezuela at 2055.
9510 R. Barquisimeto, Venezuela at 2015.
9595 R. Cullura de Bahia, Brazil at 2025.
9965 R. Lisbon, Portugal in English at 2050.
11875 R. Nacional de Nicaragua, Spanish at 2010.
12550 R. Nacional, Tanganyika at 2100.
John Adams of Sheffield returns to the column after a long absence. John has an Eddystóne EC10 receiver, a Codar PR30X preselector and a 72 foot long-wire. His log included:
9570 Radio Australia with music at 0710.
9625 Kol Israel, English news at 2035.
11620 AIR, Delhi with news in English at 2008.
11650 R. Bangladesh in English at 1715.
13220 R. Euzkadi in Basque at 2135.
15018 Voice of Vietnam, Hanoi at 2005.
21495 R. Portugal with DX News at 0800.
Nigel Knowlman of Cullompton in Devon has sent in another good $\log$ which includes:
9570 Radio Australia in English at 0830.
9912 All India Radio in English at 2000.
15105 Radio Japan, English news at 0945.
15165 R. Denmark in English and Danish at 1200.
17655 Radio Cairo, 'Arabic by Radio' at 1800.
21535 RSA, South Africa, English news at 1400.
21545 Radio Ghana in English at 1440.

Reports should arrive by the 15 th of the month and be addressed to me at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.


FIRST letter out of the postbag this month comes from GW3ZSV. He tells enthusiastically of an 80 metre contest to be run by the Sully \& District Short Wave Club on October 22. First section from $0900-1100$ and the second from 1700 to 1900. Great news-the contest is open to both licenced Amateurs and s.w.ls so there's no excuse for not having a go. Rules are just about the simplest and easiest to follow that I've ever seen. Why not send GW3ZSV a s.a.e. and get all the gen? Address is 3 Thorley Close, Cyncoed, Cardiff, South Wales. For local s.w.ls with no club at present, the Sully gang meet every Tuesday at 1930 at the Sully Bowls Social Club. Sully is about 6 miles from Cardiff on the coast road to Barry. They cater specially for the s.w.l., too.

Elated squeaks from Bill Waldron (Cwmbran) whose 144 MHz score since April 1 this year has now topped 40 counties and six countries. Bill mentions the infuriating situation during a recent 2 metre opening when London and Midland stations could be heard calling LA, SM, ON, OZ, etc., but only the faintest of whispers and bits of EU callsigns detectable down in Wales. He also remarks on hearing many QSO's of, say, Dudley talking to Wolverhampton and other local natters. He again asks why these locals don't turn their beams round towards GW. Almost every night, GW3ZTH, GW3LEW and GW3TMM are active on the s.s.b. channel. How about it you northern types-or are those beams rusted in position from the habit of local gossips?

On 80 metres, Bill complains of high static levels but still managed to log some s.s.b. from; ELOH/MM, FPOBG, OA4AS, PY2RT, PY3CGP, VE3GCS, VOlGQ/M, VP7ND, VP9BK, WA2FCA, W4NJF, W8BT, $5 Z 4 \mathrm{MO}, 8 \mathrm{P} 6 \mathrm{SJ}$ (Scout Jamboree station).
Pretty stamp and long epistle from Victor Springer (St. Lucy, Barbados) relates hair-raising tales of s.s.b. received on a receiver using battery valves and a 90 V h.t. line. Modifications include items like half of a 12AU7 as an outboard tuneable b.f.o. whose signal is injected by winding a few turns of wire around the first DF96. (Transistor types will doubtless find all this way above their heads!). (No, a 12AU7 hasn't got a collector). Heard on 14 MHz with the aid of 150 ft of aluminium wire in an unspecified shape; CE3AJE, DJ1HP, DJ2WF, DJ8EG, F30X, G3KTJ, G3XJN, G400, G6GA, GB3MKT, HB9KM, HC2JX, HK5ACP, I5MLI, JA1BRK, JA4YAR, OA4AEH, PY1AGO, VXVK2SG, VK3JW, VK9JW, XE1LAG, YK1AA, ZD3A, ZK2DX, 4Z4JX, 5V2AT, 5V4GE, 5V8AR, 7X7Y.

Chance to get yourself a nice QSL from the land of GC. Ken, GC3GPL (ex G3GPL) is now active on c.w. between $14 \cdot 000 \mathrm{MHz}$ and $14 \cdot 050 \mathrm{MHz}$ most evenings between 1900 and 2230 plus Sunday mornings. He has put up a special antenna which, he hopes, will give reliable short skip signals into the UK. He will QSL any s.w.l. who sends him a useful report on

# THE AMATEUR BANDS David Gibson, G3JDG <br> <br> Frequencies in kłz - Times in GMT 

 <br> <br> Frequencies in kłz - Times in GMT}
his signals. Address for those useful (note that word, not just a "Heard you 579 pse QSL) reports is Haut du Pre, Fern Valley, St. Helier, Jersey, Channel Isles.

Patrick Funnell (Nottingham) has an 8 -element Yagi on two metres. This feeds a dual-gate m.o.s.f.e.t. converter which in turn feeds a PW Clubman receiver used as a tuneable i.f. covering from $2-4 \mathrm{MHz}$. He comments that the Clubman is a very good receiver and includes a 2 metre $\log$ which mentions stations like; OZ1EP OZ4HW, PA0KOK, PA0LSG (using n.b.f.m.), PA0PMQ, PA0TLX (n.b.f.m.) plus numerous $G$ stations. It is interesting to note the number of stations using narrow band frequency modulation (n.b.f.m.).

Log from the quill of Alan Smith (Lancs.) using a JR310, 60 ft . long wire plus a E-ZEE Match a.t.u. He complains of ". . . a combination of night starvation and v.f.o. tuners cramp". (Wait till you get knob nippers gout, that's when we really separate the men from the boys). Alan's $\log$ for 14 MHz reads; CE3AQW, CE3OE, CO8GL, CP1JV, EA8EN, EP2TW, FC9VN, FM7WN, FY7AE, HC4BS, HK6CBS, HP1XIS, MP4TDM, OA4LV, OA4WJ, OD5AU, OX3HV, OX3JW, PJ2CC, PJ9AD, PZ5CW, TI2AS, VE1TB, VE7BNJ, VE8RCS, VK2QM, VK3DO, VK4CZ, VK5ID, VK6FD, VK7AK, VK0RC, VP2VAM, VP7MG, VS6DO, VU2ALU, VU2BX, XE1DO, XE1EH, XE3RE, ZL1VY, ZL2CB, ZL3JD, ZL4BC, ZM2ASJ, ZM4CR, 9K2AR, 8P6BC.

DX on ten metres is a GW, claims Martin "Squiggle" whose second name I can't decipher. (It's when your own g.d.o. is d.x. you start worrying!). Treasured possessions include a 640 receiver, a 20 metre dipole and a "flimsy bit of co-ax". Mr. Squiggle comments on the d.x. stations claimed by users of simple t.r.f. receivers, headphones made of draught excluders and coat-hanger aerials. He wonders if they use an SB303 receiver as a pre-amp and just forget to mention it. (Play fair lads, if you use a 30 transistor pre-amp, you gotta say).

A $9 \mathrm{R} 59 \mathrm{DE}, 7 \mathrm{MHz}$ dipole and B. Mainwaring (Walsall) raised this lot on 21 MHz ; CE3OE, CP1HG, CR6CA, EA3VC, EA6BJ, EP2TC, JA1MTP, JA2MJ, JH1TMR, JH3ONP, OD5HF, OE6PN, PA0WF, PY1MT, SM4APD, 4Z4MO, 7 X 7 Y . Interesting to note the European stations here (many haven't been included). Many s.w.ls commented on the short skip conditions which prevailed this past month. Effects like $G$ stations on 20 metres being 5 and 9 were common. Wonder what things will be like next month, and how do you think the bands will fare this coming Winter? How about a line airing your views?

Logs, in alphabetical order please to arrive by the 15 th of the month to 12 Cross Way, Harpenden, Herts.

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| 000 | BC212L | 012 | TIP33A |
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| 1 RG | . 28 | 30015 | . 581 | EABCRO | . 321 | EL500 | . 62 | PCF808 | . 88 | U801 | . 75 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 185 | . 22 | $30 \mathrm{C17}$ | .76 | EAF42 | . 50 | EM80 | . 38 | POL,82 | $\cdot 30$ | UABC80 | . 38 |
| 174 | -18 | 30018 | . 58 | EB91 | $\cdot 10$ | EM81 | . 38 | PCL83 | $\cdot 57$ | UAF42 | . 60 |
| 384 | . 26 | 30 F 5 | . 64 | EBC33 | . 40 | EM84 | -32 | P0L84 | -34 | UBC41 | . 45 |
| 3 V 4 | . 47 | 30 FL 1 | . 65 | EBC41 | . 49 | EM87 | . 50 | PCL85 | . 88 | UBF80 | . 84 |
| 5U46: | .31 | 30 FL 12 | -69 | EBC81 | . 30 | EY51 | . 88 | PCL86 | . 38 | UBF89 | . 82 |
| 5 V 46 | . 35 | 30 FL 14 | . 68 | EBC90 | . 22 | EY86 | . 29 | PCL88 | . 68 | UCC84 | . 32 |
| SY3GT | . 34 | 30 L 1 | . 29 | EBF80 | . 32 | EY87 | -29 | PCL800 | . 89 | UCC85 | -35 |
| $5 \mathrm{Z4G}$ | -35 | 30 L 15 | $\cdot 70$ | EBF83 | -39 | EZ40 | -39 | PCL805 | .88 | UCF80 | -82 |
| 6/30L2 | - 54 | 30 LJ 7 | . 87 | EBF89 | -29 | EZ41 | -39 | PENA4 | $\cdot 77$ | UCH42 | -58 |
| 6AL5 | . 11 | 30 P 4 | -65 | ECC81 | $\cdot 17$ | E280 | . 21 | PEN36C | -70 | UCH81 | . 30 |
| 6AM6 | . 13 | 30 Pl 12 | . 69 | ECO82 | . 20 | EZ81 | . 22 | PFL200 | . 52 | UCL82 | . 82 |
| 6AQ5 | . 22 | 30P19 | -65 | ECC83 | . 35 | EZ90 | 25 | PL36 | $\cdot 48$ | ${ }^{1 T} \mathrm{CL} 83$ | - 56 |
| 6at6 | . 20 | 30 PL 1 | . 60 | ECC85 | . 84 | G230 | .34 | PL81 | . 44 | UF41 | . 52 |
| 6AU6 | . 20 | 30PL13 | . 89 | ECCS 04 | .54 | G232 | . 40 | PL8iA | . 47 | UF89 | .30 |
| 6BA6 | . 20 | 30 PL 14 | -85 | ECF80 | . 31 | KT41 | . 77 | Pl84 | . 31 | UL41 | . 53 |
| 6BE6 | . 21 | 35L6dT | .45 | ECF82 | . 26 | K T61 | -55 | PL83 | . 83 | UL84 | $\cdot 30$ |
| 6BJ6 | . 41 | 3.5 W 4 | . 28 | ECH35 | . 55 | KT66 | $\cdot 78$ | PL84 | . 30 | UM84 | -22 |
| 6 BW 7 | . 50 | $35 \mathrm{Z4G7}$ | . 25 | ECH4? | . 59 | LN319 | . 63 | PL500 | . 63 | UY41 | -89 |
| 6F14 | $\cdot 40$ | 50CD6G | . 68 | ECH81 | . 28 | LN329 | . 72 | PlS04 | . 63 | UY85 | . 25 |
| 6 F 23 | . 68 | 807 | . 49 | ECH83 | . 38 | LN339 | . 63 | PM84 | . 88 | VP4B | .77 |
| 6F25 | . 53 | AC/VP2 | . 77 | ECH84 | 75 | N78 | . 87 | PX25 | . 85 | W77 | -48 |
| 6J79 | . 24 | B349 | . 65 | ECL80 | .35 | P61 | . 40 | PY32 | . 58 | 277 | . 28 |
| $6 \mathrm{K7G}$ | $\cdot 12$ | B729 | . 62 | ECL82 | . 29 | PABCA | $\cdot 34$ | PY 33 | . 58 | Tranaia | rs |
| 6 K 8 G | $\cdot 17$ | CCH35 | . 67 | ECL86 | . 85 | PC86 | . 47 | PY81 | . 25 | AC107 | $\cdot 17$ |
| 6979 | . 35 | CY31 | -30 | EF39 | . 88 | PC88 | $\cdot 47$ | PY82 | . 25 | $\mathrm{ACl27}^{\text {a }}$ | $\cdot 18$ |
| 6SL7GT | . 30 | DaF91 | . 22 | EF41 | . 67 | ${ }^{\text {PC96 }}$ | -42 | PY 83 | $\cdot 26$ | AD140 | $\cdot 37$ |
| 6SN7GT | . 30 | DAF96 | .36 | LF80 | . 23 | PC97 | .36 | PY88 | . 88 | AF115 | -20 |
| 6V6a | . 28 | DF91 | . 16 | EF85 | . 28 | PC900 | . 29 | PY800 | . 84 | AF116 | . 20 |
| 6Y6GT | . 28 | DF96 | . 36 | EF86 | . 30 | PCC84 | . 29 | PY801 | $\cdot 84$ | AF117 | -20 |
| $6 \times 4$ | . 23 | DH77 | . 20 | EF89 | $\cdot 26$ | PGC85 | . 23 | R19 | $\cdot 30$ | AF125 | $\cdot 17$ |
| $6 \times 507$ | . 28 | DK32 | . 38 | EF91 | $\cdot 18$ | PCC88 | . 88 | R20 | . 70 | A F127 | $\cdot 17$ |
| 10P13 | . 53 | DK91 | . 28 | EF92 | -27 | PCC89 | $\cdot 43$ | U25 | . 73 | OC26 | . 25 |
| 12AT7 | . 17 | DK92 | .50 | EF98 | . 85 | PCC189 | . 48 | U26 | 70 | $0 \mathrm{OC4} 4$ | $\cdot 12$ |
| 12AU7 | . 20 | DK96 | 45 | EF183 | . 28 | PCC805 | . 70 | U47 | . 78 | OC45 | . 12 |
| 12AX7 | . 22 | DL35 | .40 | EF184 | . 31 | PCF80 | -28 | U49 | $\cdot 70$ | $0 \mathrm{C71}$ | -12 |
| 19BGAG | .75 | DL92 | .26 | EH90 | . 34 | PCF82 | .83 | U52 | . 81 | OC72 | $\cdot 12$ |
| 20 F 2 | . 67 | DL94 | . 47 | Elis3 | . 55 | PCF86 | . 46 | U78 | $\cdot 24$ | 0C75 | $\cdot 18$ |
| 20P3 | . 75 | DL96 | . 28 | EL3 EL41 | . 45 | PCF800 | . 58 | U19] U193 | . 68 | $0 C 81$ 0.81 | . 18 |
| 25L6GT | $\cdot 18$ | DY86 | . 24 | EL84 | . 23 | PCF802 | .40 | U251 | . 64 | OC82 | -12 |
| 25U4GT | 57 | DY87 | .24 | EL90 | . 28 | PCF805 | . 58 | U301 | . 88 | OC82D | $\cdot 12$ |
| 30 CL | . 28 | DY 802 | . 83 | EL95 | .33 | PCF806 | .56 | U329 | . 68 | 0 Cl 70 | . 28 |

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## Virtual Earth in practice

Theory is all very well; an interesting intellectual exercise. Sooner or later, we must put our theories to the test, our principles into practice, and it is then-as any constructor will tell you-that the moment of truth comes. When you switch on, hold your breath, tentatively measure, go back meticulously over every step you have made and find the snags. That feeling of utter elation when a self-designed, self-made circuit bursts into life has no parallel in the world of module-massing. Humble though our circuitry may be, we are tempted to say: 'A poor thing, but mine own . . .'

So this month, following a discussion of virtual earth principles that may have left some of you


Fig. 61 (a): Redrawn circuil of Fig. 53 wilh some components omilted for simplicity. (b) base bias can be derived this way, included in (c) we have redrawn version of (a).
wondering what all the fuss is about, we put our theories into practice. Again, we shall proceed step by step, explaining why each component has to be its particular value, and what it does. Only this way is it possible to adapt ideas to your particular situation, i.e., to make your own design.

Part 10 in September $P W$, left us with a directly coupled pair of transistors, the first device acting as a collector-follower and with a potential divider for the derivation of its base bias. The second transistor operated in the common collector mode-that is, it was an emitter follower-and its base bias was derived from the collector of the first transistor, to which it is, (see Fig. 53, page 429, September PW) directly connected.

In the subsequent issue, under the guise of giving us all a breather, we talked about virtual earth circuitry, but you will have noted, in Fig. 60, that we again used a pair of transistors, intimately connected, directly coupled, so that their combined operation can be regarded, with some provisos, as that of a single unit.

Now let's take a look at a redrawn circuit of our earlier arrangement. Fig. 61 (a) is the redrawn circuit of our previous direct-coupled pair, this time with some components removed for simplicity. These are the base bias resistors, the input and the output coupling capacitors.
We still need base bias current, and this time we get it from the emitter of the second transistor. The important component is $R_{b}$. Take a look at another way of presenting such a circuit, Fig. 61 (b), which is really a redrawn Fig. 51 ( $P W$ Aug. 1972): now compare this with (c), which is, of course, a redrawn version of (a). The essential modification is that the base-emitter junction of the second transistor is included now in series with $\mathrm{R}_{\mathrm{b}}$. The similarity of the circuits is significant: similar spreads can be expected.

## Gain

With this circuit, just as with the circuit of Part 10. gain is determined by the ratio of collector and emitter resistors of Tr1, which is to say, Rcl/Rel.

But this gain is before any feedback is applied. It is called the OPEN LOOP GAIN, and is denoted by the term $\mathrm{G}_{\text {o. }}$. As we shall see later, though it can be calculated fairly enough, it is not quite so easy to verify by practical measurement.

When we put a bit of negative feedback in circuit, conditions alter. We are now taking a portion of the output and applying it to the input of our amplifier in such a way as to reduce the gain in proportion to the growth of the signal. If we apply the same feedback another way, of course, we can cause the amplifier to increase its gain in accordance with the signal, and thus go haywire. We are concerned here with negative feedback, reducing the gain selectively. So we get CLOSED LOOP GAIN.

The determination now is from the value of that feedback and the source input. In plain symbols the ratio of feedback resistor to source resistance, or $\mathbf{R}_{\mathrm{b}}$

$$
\mathrm{R}_{1}+\text { source resistance }
$$

That magical term (if you read the right brochures!) the 'Virtual Earth' occupied us last month and we shall now return to it, as in Fig. 62.


Fig. 62 : Virtual earth amplifier with feedback emitter Tr2 to base of Trt.
Here, the virtual earth point is indicated and we must remember that the effective resistance at this point is determined by the ratio of that feedback resistor $\mathbf{R}_{b}$ and the open loop gain of the stage. This is $\mathbf{R}_{\mathrm{b}} / \mathbf{G}_{\text {o }}$. Take a close look at this circuit, study it so that it can be recognised immediately, for this virtual earth circuitry will not show $\mathbf{R}_{\mathrm{r}}$, even dotted, as we have, and it is very easy to miss the implication. So let's now start to calculate the component values, for a practical circuit that can be part of a mixer, a microphone pre-amplifier, a variable gain pre-amplifier for an oscilloscope, and several other things.

## Calculations

In Fig. 63(a) we have the same circuit, again shorn of unnecessary frills, and now we have a 'rail voltage', or h.t. positive, of Vcc, whichever term your age group has committed you to using, of 20 volts. The point is that we want a reasonable voltage 'swing' and if a collector is going to go up and down like a yo-yo, there will be a need for a high enough Vcc to accommodate this change in collector voltage. Thus far, we have been managing with 9 -volt
batteries, convenient, easily obtainable, a constructor's aid. Now we must face the need for adding several in series or building a power supply operated from the a.c. mains. We shall look at the latter in a moment. For now, remember that the Vcc is 20 volts and for an emitter current of 2 mA in the $\operatorname{Tr} 2$ circuitry, and a voltage at the emitter of about half the supply, we get $\operatorname{Re} 2=10 / 0 \cdot 002=5,000$ ohms.
Remembering that the base-emitter voltage will normally be 0.6 V for the BC109, we say the base voltage will be 10.6 V , and this is also the voltage at the Tr collector. We have already noted that for good stability it is desirable that the voltage drop across $\mathbf{R b}$ is small. Under these proposed conditions, Tr 1 would be operating with a collector to emitter voltage of a mere 600 mV .
Now it is perfectly possible to use a transistor under these conditions, but the signal swing will be somewhat limited, so we look for an alternative way of doing things. One method is to split the feedback point, as in Fig. 63(b). Rel is maintained at the same ohmic value, but the feedback is tapped off


Fig. 63 (a): V.E. circuit redrawn. (b) tapping off of feedback point.
from the junction of the divided Re2, so that 5 volts is available at the base of Trl as bias. The voltages at $\operatorname{Tr} 2$ base and collector have not altered, so we have the condition now where Trl operates at a col-lector-to-emitter voltage of something like 5.5 volts.
Again allowing a 0.6 volts difference between base and emitter, we can now calculate Trl emitter resistor. The current flowing through Rel can be large, using a small value of resistor, but this will not help us guard against noise. On the other hand, if we aim at a very low current, and hence a large resistor, our open-loop gain goes for a chop! Take an example: if the emitter current of the first transistor were to be 100 microamps, then Rel would be $48 \mathrm{k} \Omega$ and Rcl would have to be something like $94 \mathrm{k} \Omega$. From our earlier formula, $\mathrm{G}_{\mathrm{o}}=\frac{\mathrm{R}_{\mathrm{c}}}{\mathrm{R}_{\mathrm{e}}}=\frac{94 \mathrm{k} \Omega}{48 \mathrm{k} \Omega}$, i.e., an open loop gain of something like two

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We can, true, bypass the first emitter resistor with a whacking great capacitor, but let's not complicate matters at this stage. I hope the point is gradually being made that in this type of deceptively simple transistor circuit design one has to consider both a.c. and d.c. conditions; one can affect the other. What may be fine from the bias point of view may not be the best for signal conditions. We must press on . . .

## In practice

The practical answer to the problem is to use the potential divider again, but this time as shown in Fig. 64. This circuit is more like the real thing, as our airy-fairy calculated values of components now give way to preferred values, such as $4 \cdot 7 \mathrm{k} \Omega$ for Re 2 .
Calculating the rest of the values requires recourse to the graphs for the particular transistor. If you have been following this series-and our mailbag shows, bless you, that quite a few have--you will be able to turn up the June issue and refer to Fig. 42, which was the $h_{P E}$ against $I_{c}$ curve for a BC109. We ignore the illustrative co-ordinates this time, and find that for an $\mathrm{I}_{\mathrm{c}}$ of 2 mA , the $\mathrm{h}_{\mathrm{PE}}$ is 440 , so our base current of $\operatorname{Tr} 2$ is $\frac{2 \mathrm{~mA}}{440}$ or around 4.5 microamps.


Fig. 64 : Working circuit with component values.
As this current is flowing through the collector load of the first transistor, and is very small in comparison with the collector current of Trl, we can safely disregard it in our next calculation. The emitter voltage of $\operatorname{Tr} 2$, we have said, is 10 V , and the base will be 0.6 V greater, so this 10.6 volts will be the collector voltage of Trl. Therefore, the voltage across Rcl will be the difference between the supply of 20 volts and this $10 \cdot 6 \mathrm{~V}$, i.e., $9 \cdot 4 \mathrm{~V}$.

From what we have already said about the choice between noise figures and signal swing at $\operatorname{Tr} 1$, we realise that there are limitations, upper and lower, for the emitter current, and for the balance between the feedback resistor and R2, now part of the bias circuit. Practically, our limits are from a minimum of 3 microamps and an upper of 200 microamps, which gives us R1 + R2 of between $47 \mathrm{k} \Omega$ and $3 \cdot 4 \mathrm{M} \Omega$.

Plenty of choice? Maybe, but for a compromise value of emitter current that gives us preferred values of resistors and practical conditions, we have plumped for a figure of 80 microamps, as shown in

Fig. 64. So $\mathrm{R}_{\mathrm{c}}$ becomes $\frac{9.4}{80 \times 10^{-6}}$ or approx. $120 \mathrm{k} \Omega$.
An open loop gain $G_{o}$ of 100 is a healthy figure and as this is $\frac{\mathrm{R}_{\mathrm{c}}}{\mathrm{R}_{\mathrm{e}}}$ so $\mathrm{R}_{\mathrm{e}}$ becomes $\frac{120 \mathrm{k} \Omega}{100}$ or $1 \cdot 2 \mathrm{k} \Omega$.
If we again ignore the very small base current of the transistor, we can take it that the emitter current of $\operatorname{Tr} 1$ will be very near 80 microamps, and the voltage drop across Re will be 0.096 V , which gives us a base voltage of nearly 0.7 V . Referring again to the graph, we find for an Ic of 80 microamps we get an $h_{\text {FE }}$ of around 280 . So the base current will be $\frac{1_{\mathrm{c}} 1}{\mathrm{~h}_{\mathrm{FE}} 1}=\frac{80 \text { microamps }}{280}$ or 0.2857 microamps.

If the base current is to be provided by the potential divider R1 and R2, provided the current through these resistors is much greater than this, then Ibl can be left out of subsequent calculations. But too low a combination of these resistors will affect the bias conditions of $\operatorname{Tr} 2$ (since we shall be shunting Re2 with the R1+R2 network). If we aim at a sum of the two at least 10 times that of Re2, we shall be safe. This summed resistance would be $47 \mathrm{k} \Omega$, and the current through them would be 200 microamps approximately-nearly a thousand times Ibl.

As to those limits: the minimum current we can allow through R1+R2 will be ten times Ib1, around 3 microamps, and the sum resistance would be $3.4 \mathrm{M} \Omega$. We have a very wide choice of values, but experience tells us that we cannot dip indiscriminately into the spares box.

## TO BE CONTINUED

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FIRST, we would like to apologise if anyone has written to us and not received a reply but from correspondence we are now receiving, it appears as though a few letters have gone astray. If, therefore, you have not heard from us and would like to repeat your letters, we will be pleased to reply as soon as possible.

We have received quite a large number of "points from the post" that we feel merit publishing almost as they have been written, so, without any more ado, let's begin . . .
R. L. Graper, 16 St. John's Avenue, Chelmsford, Essex tells us...
"In June 1922 I bought the first issue of Popular Wireless . . . No. 1, Vol. 1, and later, also the first issue of Television, No. 1, Vol. 1 in March 1928. I kept them and still have them (a little soiled) in the hope that they might later become valuable as "collectors items". As the address of Popular Wireless is the same as that of your present premises, I am wondering whether you took over the above journal.

There is much of interest in this first issue, and I quote you a few items as follows.
"Have you a wireless? First take out a licence. In a week or so this may be obtained direct from any post office. At present you must apply to the Secretary of the G.P.O. Cost of licence . . . 10/-.

Article. "Mr. Selfridge expresses his views. The business man and radio, by G. Selfridge."

I do not think the radiophone will ever supercede the telephone. Many wild predictions have been made on this point, and it has even been said that before long the man in the street will have his pocket transmitter with which to call up his friends or business colleagues. That prediction is very far fetched indeed. The pocket radiophone will never replace the telephone, because it will always be limited by the number of messages that can travel through the ether at the same time. At the moment, American amateurs are realising the bad effects of unlimited radio transmission . . . The ether is crammed with their messages, which consequently get so mixed up that it is impossible to hear a coherent sentence when listening in".

Short NEWS FLASH . . . Mr. Marconi's voyage . . . Mr. Marconi proposes to carry out experiments on the Atlantic with direction finders on the Short-wave, and the Long-wave transmission. Besides his other experiments, Mr. Marconi will carry out tests for the Meteorological Office. The "Electra", Mr. Mar-
coni's steam yacht of 700 tons, is the finest equipped floating wireless laboratory in the world."

Interesting prices . . . Instrument wires. Aerial Wire $7 / 23$, which was of course seven strands of 23 swg. copper wire. Price: 6/6 per 100 feet. Double silk covered wire $12 \mathrm{swg}: 4 /-$ per lb, Single valve Detector panel: $32 /$-, single valve h.f. panel: $35 /-$. These two panels together gave, it was claimed by the advertiser, a complete two valve receiving set. This was capable of receiving all music, speech etc. from all British stations.

In my copy of Television No. 1, March 1928, there are many interesting items, such as the early experiments by Baird, with a short paragraph giving details of his first demonstration before the Royal Institution.

There are pages of advertisements for (guess what) Radio sets and parts for the radio constructor. As no Television receiver is shown or advertised, I presume this came a few months later. The first experimental transmissions on the 30 line system could be received on home made apparatus, and there are articles (most detailed) on Disc Receivers, Light sensitive cells, a cartoon drawn specially by Heath Robinson, and several articles on the future miracles of this new science . . . which indeed have in some cases become fact today.

In this year there were quite a number of Valve Portables, then priced in the $£ 30$ region, but one would rather call them "transportables" I think.

## Bintage $\mathbb{E}$ quipment

Mr. G. H. Wallace was recently presented with a pair of old headphones and he would like to know what impedance they are and what vintage they are likely to be.

continued on page 641

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 $/ 450 \mathrm{~V} \cdots 14 \mathrm{p} \quad 1000 / 25 \mathrm{~V} \cdots \quad 8 \mathrm{p}) 30+82 / 250 \mathrm{~V}$ 

$6 / 450 \mathrm{~V}$ \& 1 bp \& $1000 / 50 \mathrm{~V}$. \& 47 p \& $32+32 / 450 \mathrm{~V}$ <br>
$82 / 450 \mathrm{~V}$ \& 80 p \& $8+8 / 450 \mathrm{~V}$ \& 18 p \& $850+50 / 825 \mathrm{~V}$
\end{tabular}




2, 4, 5, 8, 1f, 25, 80, $50,100,200 \mathrm{mF}$. 15 V .10 p
 2000 mP . 6 V .25
 CERAMIC 1pF to $0.01 \mathrm{mF}, 4 \mathrm{p}$. Silver Witiof $5000 \mathrm{pF}, 4 \mathrm{p}$ PAPER 850V-0.1 4p, $0.513 \mathrm{p} ; 1 \mathrm{mF} 15 \mathrm{p}$; $2 \mathrm{mF} 150 \mathrm{~V} 150^{\circ}$. 500V-0.001 to $0.054 \mathrm{p} ; 0.15 \mathrm{p} ; 0.258 \mathrm{p} ; 0.4725 \mathrm{p}$ SIL VER MICA. Clone tolerance $1 \%$ 2-2-500p $8 \mathrm{p} ; 500-2.200$ pF 10p; 2,700-5,600pF 20p; 8,800pF-0.01, mid 30p; each TWIN GANG. " $0-0$ " $808 \mathrm{pFF}+178 \mathrm{pF}, 65 \mathrm{p}$; SIow motion drive $865+886$ with $25+25 \mathrm{pF}, 50 \mathrm{p} 500 \mathrm{pF}$ alow motion, tandard 5p; amsll 8-atane 500pF $21-60$.
SHORT WAVE, SINGLE. 10 pF 80p; 25pF 55p: 50 pF 55p NEON PANEL INDICATORS 250V AC/DC Amber 20p REsir HIGH STABILITY. $W$. $2 \% 10$ ohma to 1 meg., 10 p . Ditto $6 \%$ Preferred valaes 10 ohms to 10 meg., 4 p.
WIRE-WOUND RESISTORS 5 watt, 10 watt, 16 watt 10 ohms to $100 \mathrm{E}, 10 \mathrm{p}$ each; 2 F watt, 1 ohm to $8 \cdot 2 \mathrm{ohms} 10 \mathrm{D}$

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## 

METAL PLINTH \& PLASTIC COVER Cut out for most Garrard or
E.S. Will play with cover in position. Latest design. Covered in black ieatherefte.

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 $350-0-85080 \mathrm{~mA} .6 .8$ v. $9.6 \mathrm{a} .8 .8 \mathrm{\nabla} .1 \mathrm{a}$. or 5 v .2 a . 83.00 $300-0-300$ ₹. $120 \mathrm{~mA}, 8.3$ v. 4 a. C.T.; 6.8 ₹. 2 a a., MINIATURE $200 \nabla .20 \mathrm{~mA} . \mathrm{B}_{8} 8 \mathrm{~B} .1 \mathrm{~s} .24 \times 24 \times 2 \mathrm{in}$,
 HEATER TRANS. $63 \% .3 \mathrm{a}$.
Ditto tapped sec. 1.4 च., $8,8,4,5,6-3$ v. 14 amp GENERAL PURPOSE LOW VOLTAGE. Tapped OU 05 p at 2 amp. $8,4,5,6,8,9,10,12,15,18,84$ and 80 , 19.25 $1 \mathrm{zmp} .8,8,10,12,16,18,20,24,30,36,40,48,80,42 \cdot 30$
$2 \mathrm{amp} .6,8,10,12,16,18,20,84,30,36,40,48,8,83-25$ 5 mmp . 6, 8, 10, 18, 18, 18, $20,24,80,86,40,48,00.2575$ ADTO TRANSFORMERS 1157 . to 2307 . or 280 v . to 115 CEAR GER TRANSFORMERS. Input $800 / 250$ CEAR GER TRANSFOR 1
 8 or 12v. outpots, $1+$ amp. $40 \mathrm{p} ; 2 \mathrm{smp}$. 55p: 4 smp .85 p LUCAS \&DS500 Bridge 70\% 5 amp El .
MAINS ISOLATING TRANSFORMER. Primary 0-110 2407. Secondary 0240 v .3 amps . 780 watts. Insulated ase with flxing feet. Famous make
BARGAIN \& 0
$1 \%$ inch DIAMETER WAVE-CHANGE SWITCHES 25D
2 p. 2 -way, or 2 p. 6-way or 3 p. 4 -way 25p each. 1 p. 12 -way 2 p. 4 -wey 25
TOGGLE SWITCHES, 1 p . 14p; dp. 18p; dp. dt. 23p.
"THE INSTANT" BULK TAPE ERASER \& HEAD DEMAGNETISER 200/250v. A.C. $\mathbf{L 2} \mathbf{2 5}$ Poiflet S.A.E.

E.M.I. TAPE MOTORSPoot 15 p.

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120 v . or 240 v . A.c. $2,400 \mathrm{rpm}$. 2 -pole


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Trimmers without Xial-60p per doz. plus 17in p.p.
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\begin{array}{lcccr}
19^{*} / 21^{\prime \prime} \text { slimline }\left(10^{\circ}\right. \text { tube) } & . & \mathbf{2 5 - 0 0} \\
23^{*} \text { slimline } & \ldots & \cdots & \cdots & \mathbf{8 7 . 5 0} \\
19^{*} \text { B BC } 2 \text { sets } & \cdots & \cdots & \cdots & 214.60
\end{array}
$$

## TUBES EX EQUIPMENT (Tested)

 SINGLE PANEL$$
\begin{array}{ccc}
19^{*} / 21^{\circ} \text { any type } & \cdots & . \\
23^{\circ} \text { any type } & \cdots & \cdots \\
\text { TWIN PANEL } & \cdots \\
19^{\circ} \text { bonded } & \cdots & \cdots \\
\hline
\end{array}
$$

All tubes add $\& 1$ cartiage.
VALVES EX EQUIPMENT

| - Eb91* | 5p | 30L15 | 124p | PL36 | 281 p |
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| ECC82 | 12\#p | PC97 | 17tp | PY81 | 15p |
| EC180 | $7 \pm 7$ | PCF86 | 171p | PY800 | 15p |
| EF80 | 124p | PC84 | $71 p$ | PY82 | 719 |
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For Ferguson 850900 chassis. Adaptable for most UHF Chasein at-60, p. \& p. 50p.

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Smiths reconditioned switchmaster MK III. Decimallzed. Perfect working order. 12 for $\mathbf{5 2 5}$ delivered. For sample send $52.50 \mathrm{c} . \mathrm{w} .0$.

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GARtARD mag tape decks: $17 \mathrm{ips}, 50 \mathrm{v}$ golenold operated brakes etc., Mains voltage Hotore 87.50 each, pp $21 \cdot 23$.
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ELECTRIC MOTORS, HOOVER OR CROMPTON PARKINSON. 250 r . single Phase A.C.
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LEVER ACTION P. 0.1000 TYPE SWITCHES.
Lock 4-pole changewer 15p, ip $3 \leqslant \mathrm{p}$. (ex-equip.) Lock 2 Pole Cbangeover 10p pp 31p. (ex-equip.)
aUdio leads
Screened Phono Leads 4i" long, 15p
3.5 mm JACK/3. 5 mm JACK $\mathrm{i}^{\prime} \mathrm{G}^{\text {² }}$ long 40p

6-Pin Din A Type, 5-PIN A TYPE. Approx $5^{\circ} \mathrm{lnng}$
70p. pp above itenis $5 \frac{1}{2} \bar{p}$.


RUBBER 3 P1N 5 AMP NON REVERSIBLE CABLE 0p pp 5ip.
SOLENOIDS 18 VOLT PULL ACTION
$2^{*} \times 1^{\prime \prime} \times \mathbf{1 " ~}^{40 \mathrm{p}} \mathrm{pp} \mathrm{sp}$
SIEMENS MINIATURE RELAY. Double pole changeover dust eover/base 48 v 250050 p pp 6p new. OMRON MK2 MIDGET POWER RELAY, I2v DCX Double pole Changeover. New. 70p pp 5 p .
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GAPDNER'S POTTED TRANSFORMER 0-250v Input: $18 v 500 \mathrm{~m} / \mathrm{n} 50 \mathrm{y} 150 \mathrm{~m} / \mathrm{a}$, fiv $250 \mathrm{~m} / \mathrm{a}$ Output Size $3^{\prime \prime} \times 2 z^{\prime \prime} \times 22^{\prime \prime}, 21 \cdot 00$, pp 20 p. Ex equip tested
TELESCOPIC AERIALS
Chromed $7^{\circ}$ closed $28^{\circ}$ extended 6 section ball jointed base 23p pp 8p new.
MOLLARD 4 DM 160 INDICATORS in plastic holder/cover ex-equip. size approx. $1 \mathbf{z}^{*} \times \mathbf{1 t}^{+\times} \times \mathbf{1}^{-}$ 88p pp 8p.
PRINTED CIRCUIT BOARD/19 ACY 19'S 10 OA 200
Diodes: 1 reed relay: 1 AZ 229 zenuer ass. capacitor/ resistors. Power supply $22 \mathrm{v} 250 \mathrm{~m} / \mathrm{A}$ DC. Output 240 v . AC 81 pp 20 p ex-equip.
TOGGLE SWITCHES. Single pole Double Throw ex-equip. new condition. 50p doz. pp 13p.
PAINTON PLUG SOCKETS Type 159 series working voltage 350 y AC/DC current max. 3 amp AC/DC el pp fip. 31 WAY PLUA \& SOCKET. E1-50. $\begin{array}{ll}\text { el } & \mathrm{pp} \\ \mathrm{pp} & 6 \mathrm{p} .\end{array}$

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Adjacent Elstree Mainline Station Tel: 01-953 6009.

## GOING BACK -continued from page 638

They are about $2^{1}{ }_{8}$ in. diameter and constructed from aluminium. We have persuaded the great "PAX" of "Practically Wireless" fame to switch off his vintage Hawaiian 78's for a moment and do a sketch to show how Mr. Wallace's headphones look.
If you have any information on these, please send it to Mr. Wallace at 3 Woodside Avenue, Kilmarnock, Ayrshire, Scotland.

Miss C. Crabb writing from Bournemouth, Hampshire. tells us that she recently unearthed the receiver and speaker shown in the photographs.
The receiver was licensed under the Marconi patent and has reactance tuning. The grid bias battery is fitted inside the cabinet.

Miss C. Crabb's horn loudspeaker


Miss Crabb asks if this equipment is of any value as a curio or as a collectors' item.

In answer, we can only stress once more that equipment of this vintage (and indeed any equipment pre-1930) is certainly worth hanging on to. Collectors would put a value on it and even the person wanting such an item as just a "curio to put on the shelving in the lounge" would possibly be prepared to offer a reasonable sum.


If any other readers have any information on this receiver and would care to write to us, we will be pleased to pass on the information to Miss Crabb.

## Zintage 出etord 進ajaar

We would like to thank readers who have written in letting us know of their old gramophone records and equipment. We hope to run an article in a future issue on early gramophones, and 'will find the information you have provided will help us greatly in our project.

Readers may also be interested to know that the 5th annual Vintage Record Bazaar will be held on Saturday, Oct. 14th, at St. Silas Hall, 74 Penton Street, Islington, London, N.l.

The first two of these record sales, five years ago were held in conjunction with vintage car rallies, there being 15 record stalls.

This year, there will be 50 stalls with over 30,000 vintage records on sale. Vintage post cards will also be in evidence if you are interested in collecting them. In addition, there is a strong possibility that there will be some vintage radio equipment there!

The Bazaar is open from $10 \mathrm{a} . \mathrm{m}$. to $5 \mathrm{p} . \mathrm{m}$. and admission is 25 p.

If anyone would like any further information on the Vintage Record Bazaar, phone John Carter (Shurlock Row) 539, or if you are phoning from London, 0735-81-539.

## ELECTRONOTES-continued from page 625

the physical properties of the medium at various depths.

These points are important since the acoustic signal pulsed out by the transmitter in an "active" system could be diffracted in a curved path and thus there would be certain completely blind areas.

The thermal profile of the water at various depths can be quite a complex thing. This profile will alter daily being dependant upon the day to day variation of the solar activity of the sun. Again, winds and rain cause a mixing action which is yet another variable.

We have barely scratched the surface of this fascinating subject. However, it appears that we have a long time to wait before some boffin comes up with a useful sonar device for the accurate location of the soap in murky bath water. Or does some reader just happen to have a practical circuit handy?


## I.C. Stabilised Power Supplies-continued from page 625 NOTES

Although component layout is unimportant two details must be observed. The voltage across Cl must be less than 36 V before the i.c. is inserted into the circuit. If the voltage across Cl exceeds 36 V this voltage can be reduced by altering the tapping of the transformer. Secondly it should be noted that the output power transistor will, under current limiting conditions, have to dissipate about 40W. An adequate heat sink for this component must be provided, the chassis of the supply being used for this

## components list

R1 $2 \mathrm{k} \Omega \frac{1}{2} \mathrm{~W} 5 \% \quad$ R2 $560 \Omega \frac{1}{2} \mathrm{~W} 5 \%$ R3 $0.5 \Omega 3 \mathrm{~W}$ wirewound R4 $1.5 \mathrm{k} \Omega \frac{1}{2} \mathrm{~W} 5 \%$ VR1 $10 \mathrm{k} \Omega$ skeleton preset potentiometer C1, C2 $2500 \mu \mathrm{~F} 40 \mathrm{VW}$ minimum. C3 $0.01 \mu \mathrm{~F}$<br>Tr1 2N3055 , Tr2 2N3704. Tr3 2N3704<br>D1-4 3A silicon rectifiers or bridge<br>D5 BZY88C9V1 ( 9.1 V 400 mW )<br>D6 BZY88C6V2 ( $6 \cdot 2 \mathrm{~V} 400 \mathrm{~mW}$ )<br>ICI. $\mu \mathrm{A} 741 \mathrm{C}$ T1 $230 \mathrm{~V} / 24 \mathrm{~V} 1 \cdot 5-2 \mathrm{~A}$<br>Red and black insulated terminals. Neon lamp and holder 230V. Fuses 2.5A and 1A with holders. Meter 50 VDC. Mains input plug and socket. On-off switch. Case and chassis. Veroboard $4 \frac{4}{} \times 4 \mathrm{in}$.

The components list above applies only to the variable voltage supply. The veroboard used in the prototype has pins fitted which plug into a chassis mounted socket.
purpose. The transistor is however electrically isolated from the chassis by using a greased mica washer and bushes. Although no other difficulties should arise it is recommended that the semiconductors, and especially the i.c. should be guaranteed to be to the full specification. Providing all these precautions are taken the supply should prove to be completely reliable.

## MAXWELL

by G8DSH

"I should like to introduce you to the wife of our Stereo expert!"

## CO! CO! CO! CO! CO!

BOOKS WANTED
buy or borrow any books on crystal radios, and extremely simple transistor amplifiers or radios: also any useful circuits.-B. Ewing, 25 Water Lane. Seven Kings. Ilford, Essex.

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Sunderland SR4-OBB. 1960-1970 inclusive tor sale.-E. Dale, 2 Fordham Square
.. The following volumes of P.W. Vol 42 (May 1966-A pril 1967), vol 43 (May 1967April 1968), Vol 44 (April 1968-May 1969), Vol 45 (May 1969-April 1970) at 75p pet vol postage included.-C. Butler, 81 Rutland Avenue, Southend-on-Sea. Essex.
vol Practical Wireless $1948-1965$ and Practical Television 1950-1964. Offers to $J$ Skelly. 8 Sylvia Avenue, Hulton. Brentwood, Essex
fuit issues for 1967-1968-1969-1970 and 1971 in mint condition, for which I will accept the best offers, plus postage--L. J. Bellingham, Woodside, Red Lane, LImpfield, Surrey.
W. 1962-1964 (inc) Feb 1962, May 1963, Sept. 1964 missing. P.E. Nov 1964Apr, 1966 with Vol. 1 binder. Offers to C.J. Barrett, $13 B$ Millers Road, Brighton, BN1 5 NP .
M. Wolock, October Cottage, Station Road Ell in good cond
..P.E. from Oct 1969 -Dec. 1971, sell at cover price \& $P$ \& $P$, any reasonable offers accepted.-S. Smith, Ardvaros, Main Street, Polmont, Falklick.
. many P.W. and P.E, from 1964 on.-A.G. Hodgett, 3 Ely Way, Feligate, Co Durham.
.approx 85 P.W. 1962-1969 also most blueprInts. Few P.W., 1949-1950, 1958, 1960, 1961. 17 In ali. approx 100 PT 1952-1963. Most Short Wave Mag', 1963-1967 Offers.-M. Blackburn. 107 St. Paul's Road, Jarrow, Co. Durham.
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.various dates: P.Ts. 1959-1965, P.Ws. 1968-1971 P.Es 1965-1968. P.W. 21st ANN; Oct. 1953, (1) "Wireless Constructor" 1931.-S.W. Owen Bryngwynedd, Rhiw.
 1970. Also P.W. April-Dec. 1968.-V. Simpson, 107 Derryhale Road, Portadown, Co. Armagh, NI.

I have the following issues for disposal Feb, Apr-Jun. Aug, Oct, Nov. 1967, Mar-Jun, Aug. Sep. Nov, Dec. 1968 , Jan., Feb., May 1969, Feb.-May, Oct. 1970 Aug. 1971 and 2 each Jul. 1967. De6B of P.W. Jun. 1966. Jun. 1967, Dec.. 1967, Jan., Apr.Aug. Nov., Dec. 1968, Mar. 1969, Feb., Mar., Nov. 1970, Feb. Mar., Aug. 971 2 each Jul., Aug. Mar,-Nov. 1971 Reath, Sussex. RH16 3 LY. 1960 Radlo Communication from 1968. Wireless World1959/1960/1961. S. A.E. list.-R. Forsberg, 123 Harestone HIII, Caterham, Surrey, CR3
6DL.P.W. 1958 onwards. Rc 1951-1957 Bound, also RC 1958-Sept. 1965, and Dec 1968 Aug. 1970.-G. F. Allan, 21 Borrowdale Close, Redbridge. Ilford.
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## CORRESPONDENTS WANTED

.wanted to correspond; Someone of my own age (17). Interested in Electronics and Modern Music.-P. Richards, 7 Holwood Drive. Whalley Range, Manchester M16 8 WS. ..girl or boy of my own age (14) to correspond or tapespond with.
and
3 ave a Prinzsound TR-5 $17 / 8$ cassette recorder and a Ferguson 37 and
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| AAZ13 | 10 p | BY100 | 15p | 0A47 | 8 p |
| AAZ15 | 10p | BY103 | 88p | OA70 | 78 |
| BA100 | 155 | BY122 | 87 | OA73 | 10 p |
| BA102 | 80 p | BY124 | 15p | OA79 | 7 |
| BA110 | 95 | BY126 | 12\% | 0481 | 8 P |
| BAll1 | 27 p | BY127 | 15p | OA85 | 7p |
| BA112 | 70p | BY164 | 68 p | OA90 | 7 p |
| BA115 | 7p | BY210 | 85p | OA91 | 7 p |
| BA141 | 32 p | BYZ11 | 30 p | OA95 | 7p |
| BA142 | 88 p | BYZ12 | 807 | OA200 | 7p |
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# Sinclair Project 60 

## Stereo 60



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## pre-amplifier/control unit

The versatility of Project 60 high-fidelity modules is well demonstrated in this excellent unit. It provides the facilities essential to good stereo and will enhance the quality of any system it is used with, whether Project 60 or any any other top line power amplifiers. Compact, yet robustly constructed, the unit is easily panel mounted and will operate satisfactorily from 18 to 35 volts supply. Silicon epitaxial transistors are used throughout to achieve a very high signal to noise ratio with excellent separation between channels. Distortion at maximum output is barely $0.02 \%$ with magnetic p.u. input. Accurate equalisation is provided for all inputs, which are selected by push buttons. For maximum effectiveness, the Sinclair A.F.U. is recommended for use with the Stereo 60 pre-amp/control unit. A comprehensive manual supplied with Project 60 modules makes installing and connecting easy and ensures best possible results from your system.

SPECIFICATIONS
Input sensitivities: Radio-up to 3 mV Mag pu. 3 mV correct to R।A A curve $\mathrm{Mag} p \mathrm{u} .3 \mathrm{mV}$ correct to RlA A curve
-1 dB .20 to 25.000 Hz Ceramic p.u -up -1 dB .20 to 25.000 Hz Ceramic p.u -up
to 3 mV . Aux $-u p$ to 3 mV Output: 250 mV to 3 mV ; Aux-up to 3 mV Output: 250 mV Signal to noise ratio: better than 70 dB Channel matching: within 1 dB
Tone controls: TREBLE +12 to -12 dB at 10 KHz : BASS • 12 to -12 dB at 100 Hz Front panel : brushed aluminium with black hnobs and controls
Size: 66 40:207mm


aving introduced integrated Circuts to hus onstructors with the +C 10, the frst time an C ad ever been made avaliable for such purposes ve have followed it with an even more efficient ersion. the Suner IC 12. a most exciting advance ver our or gina unit This needs very few ex ernal resistors and capacitors to make an - astomishingly good high fidelity amplifier for use - with pick-up. FM radio or smali P A set up. etc The free 40 page manual supplied. detals many other applications which this remarkable IC make possible it is the equivalent of a 22 tran
istor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for dll requirements. The Super IC. 12 is compatible win Project 60 modules which would be used with the $Z 50$ and $Z .30$ ampl fiers Complete with free manual and printed circuil board

## SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak) 6-8 Frequency Response: 5 Hz to $100 \mathrm{KHz}+1 \mathrm{~dB}$ Total Harmonic Distortion : tess than $1 \%$. (Typical $0.1 \%$ ) at all output Less than $1 \%$. (Typical $0.1 \%$ at all output powers and frequencies in the audio band ( 28 V )
Load Impedance: 3 to 15 ohms Input Impedance: 250 Kohms nom nal Power Gain : 90 dB (1,000,000,000 times) after feedback Supply Voltage: 6 to 28 V . Quiescent current: 8 mA at 28 V . Size: $22 \times 45 \times 28 \mathrm{~mm}$ in clud ng pins and heat sink
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The $Z .30$ and $Z .50$ are of advanced design using silicon epitaxial planar transistors to provide unsurpassed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at 15 W ( $8 \Omega$ ) and all lower outputs. Whether you use $Z .30$ or $Z .50$ amplifiers in your Project 60 system will depend on personal preference, but they are the same size and are intended for use principally with other units in the Project 60 range. Their performance and design are such, however, that $Z .50$ s and $Z .30$ may be used in a far wider range of applications
SPECIFICATIONS ( $Z .50$ units are interchangeable with Z. 30 s in al/ applications). - Power Outputs Z. 3015 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 30 hms using 30 volts
Z.50 40 watts R.M.S. into 3 ohms using 40 volts 30 watts R.M.S. into 8 ohms using 50 volts

Frequency response: 30 to $300.000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$. Distortion: $002 \%$ into 8 ohms . Signal to noise ratio: better than 70 dB unweighted. Input sensitivity: 250 mV into 100 Kohms (for 15 w into $8 \Omega$ ). For speakers from 3 to 15


## Project 60 Stereo F.M. Tuner

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to notse ratio. Now. Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit colls, an 1.C. in the specially designed stero decoder and switchable squelch circuit for stlent tuming between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

$7 \mu \mathrm{~V}$ for lock-in over full deviation. Squelch $7 \mu \mathrm{~V}$ for lock-in over full deviation. Squelch level: Typıcally $20 \mu \mathrm{~V}$. Signal to noise ratio: $>65 \mathrm{~dB}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}$ ( $\pm 1 \mathrm{~dB}$ ). Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation Stereo decoder operating level: $2 \mu \mathrm{~V}$. Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R M S. maximum Dperating voltage: 25-30VDC. Indicators: Stereo on; tuning. Size: $93 \times 40 \times 207 \mathrm{~mm}$.

## A.F.U. High \& Low Pass Filter Unit

For use between Stereo 60 unit and two $Z .30$ s or $Z .50$ s. The unit is very easily mounted and is unique in that the cut-off frequencies are continuously variable. As attenuation in the rejected band is rapid ( $12 \mathrm{~dB} /$ octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system, There are two filter sections - rumble (high pass) and scratch (low pass). H. F. cut-off ( -3 dB ) variable from 28 KHz to 5 KHz . L.F. cut-off ( -3 dB ) variable from 25 Hz to 100 Hz . Distortion at 1 KHz $(35 \mathrm{~V}$, supply) $0.02 \%$ at rated output. Operating voltage from 15 to 35 V . Current 3 mA . Size: $66 \times 40 \times 90 \mathrm{~mm}$.

## Power Supply Units

Designed specifically for use with the Project 60 system of your choice. Use PZ. 5 for normal Z.30 assemblies and PZ. 6 or PZ. 8 where a stabilised supply is essential
Typical Project 60 applications

| System | The Units to use | together with | Units cost |
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| Simple battery record player | 2.30 | Crystal P.U., 12 V battery volume control, etc. | £4.48 |
| Mains powered record player | Z.30, PZ.5 | Crystal or ceramic P.U volume control, etc. | £9.45 |
| 12W. RMS comtınuous sine wave stereo amp. for average needs | $\begin{aligned} & 2 \times 2.30 \text { s, Stereo } \\ & 60 ; \text { PZ. } \end{aligned}$ | Crystal, ceramic or mag P U,FM Tuner, etc. | ¢23.90 |
| 25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers | $\begin{aligned} & 2 \times 2.30 \mathrm{~s} \text {, Stereo } \\ & 60 ; \text { PZ. } 6 \end{aligned}$ | High quality ceramic or <br> magnetıc P.U.. F.M. <br> Tuner, Tape Deck, etc. | f26.90 |
| 80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms) | $2 \times 2.50$ s, Stereo 60; PZ.8, mains transformer | As above | £34.88 |
| Indoor P.A. | 2.50, PZ.8, mains transformer | Mic. guitar, speakers, etc., controls | £19.43 |
| F.M. Stereo Tuner ( £25) \& A.F.U. (£5.98) may be added as required. |  |  |  |
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