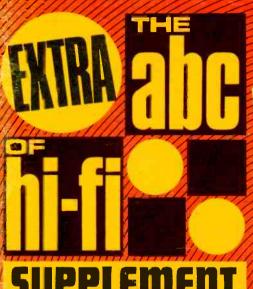
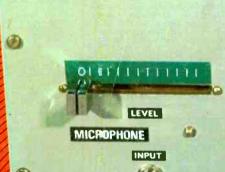
NOVEMBER 1972

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SUPPLEMENT





PAReverberation





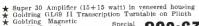
8 WATT SILICON AUDIO AMPLIFIÉR

SIMPLE RF SIGNAL GENERATOR

RSC HIGH - FIDELITY STEREO PACKAGE OFFERS Four fully wired units units and the property of the

GUPER 30 AMPLIFIER (15+15 watt) in veneered housing GARRARD 8P25 MK III Turntable on Plinth with cover GOLDRING G850 Magnetic cartridge with diamond stylus PAIR OF STANWAY II Speaker Lints Special Total Price 287-95

Terms: Deposit £12.77 and 9 monthly payments £9.39 (Total £97.28).



★ Goldring Magne ★ Goldring Magne P.U. Cartridge

P.U. Cartridge Pair of Stanway II

Special Total £98-63 Price

and of Statistics Price speaker units appeaker units s: Deposit £13.63 and 9 monthly payments £10.62



★ TA12 AMPLIFIER 6.5+6.5 watt in veneered housing

★ GARRARD SP25 MK III Player unit on Plinth

★ SONOTONE 9TAHC Ceramic Cartridge (diamond stylus)

★ PAIR OF DORCHESTER Loudspeaker Units Or Deposit £8.40 and 9 mthly.

payments of £6.32 (Total £65.28). Trans. Plastic Cover £3-15 extra.

Special Total Price Carr. £1.25.

PACKAGE AS ABOVE but with Garrard 3000 Antochanger and Sonotone 9TA Ceramic Cartridge in lieu of SP25. Carr. £1-25. £51:75

Or Deposit £7.50 and 9 monthly payments £5.64 (Total £58.26). Trans. Plastic Cover £3.15 extra.

★ Send S.A.E. for coloured brochure showing other money-saving offers.

'YORK' HIGH-FIDELITY 3 SPEAKER SYSTEM

* Moderate size only 25×14×10in. approx.

COMPLETE KIT £23

Corr. 65p

edance 15 ohms Impedance 10 0mms

**Performance comparable with units costing considerably more.

Consists of (1)12in. 15 watt Bass unit with east chassis, Roll rubber cone surround for ultra low resonance, and ceramic magnet. (2) 3-way quarter section series cross-over system (3) 8 × 5in. high flux middle range speaker. (4) High efficiency tweeter. (5) Appropriate quantity acoustic dampling material. (6) Handsone Teak veneered cabinet. (7) Circuit and full instructions. Terms: Dep. 24-60 and 9 monthly payments 22-47 (Total 226-83).

DEMONSTRATIONS AT ALL BRANCHES

HIGH FIDELITY LOUDSPEAKER UNITS

Cabinets latest style Satin Teak veneer. Acoustically lined or filled acoustic damping. Ported where appropriate. Credit terms available.

DORCHESTER (Illustrated) Size 16×11×9in. appr. Range 45·15,000 c.p.s. Rating 8-10 watts. Fitted High flux 13×8in. 49·45 Carr. 40p. Dual Cone speaker. Imp. 3 or 15 ohms.

MONARCH Size 19×10×9in. approx. Rating 10 watts. Inc. 13×8in. speaker with highly flexible P.V.C. cone surround, long throw voice coil and 10,000 line magnet. High flux pressure tweeter. Handsome cabinet. Range 35·20,000 c.p.s. Imp. 8 ohms. Gives smooth realistic sound output. See 'package offers' for Carr 50p 419·35

ance with any 8in.

Hi-Fi spkr. Size 22

× 15 × 9in.

SE12 For excellent

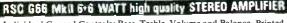
performance with 12in. Hi-Fi speaker and tweeter

R.S.C. BATTERY/MAINS CONVERSION UNITS

HI-FI SPEAKER ENCLOSURES MODERN DESIGN Teak veneer finish. Acoustically lined. Sizes approx. Carr. 35p. per enc. JESS Size 16 × 11 × 9in. SES For optimum perform-Pressurised. Gives pleasing ance with any 8in. Hi-Fi speaker. Library 15 × 9in. With any 45.35 Hi-Fi spkr. Size 22 46.47

with 10in. Hi-Fi spkr. Size 24 × 15 to in P'td.

Hi-Fi speaker and F7-87 Size 25 × 16 × 10½in. £7-87



Individual Ganged Controls: Bass, Treble, Volume and Balance. Printed circuit construction employing 10 Transistors plus Diodes. Output rating I.H.F.M. Frequency range 20-20,000 c.p.s. Bass Control ± 12db.

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Standard 200-250V. A.C. mains operation.
Attractive Black and Silver finished metal facia plate and matching control knobs.
COMPLETE KIT OF PARTS INCLUDING FULLY WIRED PRINTED CIRCUIT and comprehensive wiring diagram and instructions

11-50 Carr.
40p £11.50 Carr.

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AUDIOTRINE HI-FI SPEAKER SYSTEMS

onsisting of matched 12in. 11,000 line 15 Watt 15 ohm Consisting of matched 12in. 11,000 line is wart is omm bigh quality speaker, cross-over unit and tweeter. Smooth response and extended frequency range ensure surprisingly realistic reproduction. OR SENIOR 15 WATT INCLUDING HF126 15,000 LINE SPEAKER

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Heavy construction. Latest high efficiency ceramic magnets Plasticised Cone surround. "D" indicates Tweeter Cone providing 15,000 c.p.s. Impedance 3 or

A STATE CHOICE

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FANE 807T HIGH FIDELITY SPEAKER

A full range 8in. 10 watt unit for excellent sound quality, in suitable enclosure. Cast chassis Roll F.V.C. cone surround and long throw voice coil to achieve very low fundamental resonance of 30 c.ps. Tweeter cone is fitted to extend high note response. Frequency range 25 Hz to 15 KHz. Gauss 10,000. Impedance 3 or 8-15 \(\Omega\$. Cause 10,000. The dates 3 or 8-15 \(\Omega\$. £3.85



TYPE BMI. An all-dry battery eliminator. Size 5½×4½×2in.

approx. Completely replaces batteries supplying 1.5v and 90v, to battery radio where A.C. mains 200/230v. 50c/s is available.

COMPLETE KIT 43-25

WITH DIAGRAM 43-25

FOR USE R.S.C. TA6 6 Watt HI-FI AMPLIFIER

SE10 For outstanding results

200-250v. AC mains operated. Frequency Response 30-20,000 c.p.s. —2dB. Harmonic Distortion 0-3% at 1,000 c.p.s. Separate Base and Treble 'lilit and cut' controls. 3 input sockets for Mike, Gram, Radio or Tape. Input selector switch. Output for 3-15 ohm spkrs. Max. sensitivity 5mV Output raing I.H.F.M. Fully enclosed enamelled case, 91×21×51in. Attractive brushed silver fluish facia plate 102×31in. and matching knobs. Complete kit of parts with full wiring diagrams and instructions.

OR FACTORY BUILT WITH 12 MONTHS' GUARANTEE £10.95



£7.75 Carr. 40p.

R.S.C. TRANSFORMERS, L.F. CHOKES & RECTIFIERS FULLY GUARANTEED. Impregnated and Interleaved where necessary. MIDGET CLAMPED TYPE $2\frac{1}{4} \times 2\frac{1}{4} \times 2\frac{1}{4}$ in.

450-0-450v. 250mA, 6.3v. 4a., c.t., 5v. 3a. £5.50

FILAMENT OF TRANSISTOR POWER PACK Types 6.3:: 1-5a. 49p; 6-3v. 2a. 54p; 6-3v. 3a. 78p; 0-3v. 4a. 30; 12v. 1a. 55p; 12v. 3a. 07 24v. 1-5a.21-35; 0-18v 12a.21-10; 0,12-25-42v 2a.21-76...

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All 6/12v. D.C. output. Max. A.C. input 18v. 1a, 25p. 2a, 35p. 3a, 50p. 4a, 65p. 6a, 80p.

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150mA, 7-10H, 250 Ω 70p; 100mA, 10H, 200 Ω 60p; 80mA, 10H, 350 Ω 50p; 60mA, 10H, 50p; 60m. 400 Ω 25p.

R.S.C. MK III SUPER 30 HIGH FIDELITY STEREO AMPLIFIER

A completely new design further improved in both appearance and performance. Representing value far higher than the prices suggest.



* SATIN SILVER METAL FACIA with black letter-

* SATIN SILVER METAL FACIA with black letterIng. Black edged knobs with bright silver centres.
* PUSH-BUTTON SELECTOR SWITCHING
* PUSH-BUTTON SELECTOR SWITCHING
* ALE ON INDICATOR
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R.S.C. COLUMN SPEAKERS IDEAL FOR VOCALISTS

TYPE C482 25-30 WATTS

Fitted four 8' high flux 8 watt speakers
Overall size approx. 48×10×5in. £17.75

Terms: Dep. 43 and 9 monthly
payments £2 (Total £21)

Carr. 50p.

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Fitted four 12" 11,000 line 15 watt
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ments £3-37 (Total £34-38) Carr. 75p

Terms: Dep. 44 and 9 monthly payments £3-37 (Total £34-38) Carr. 75p

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Hi-FI AMPLIFIER

Stable 30-0,000 c/s. All benefit of cutput. Hum level -70dB. Response £786 EF86, EC083, EL34, EL34, 6Z34. Beparate Bass and Treble Controls. Sensitivity 36 millivoits. For High Impedance microphones. For Clubs, Schools, Theatres, Dance Halls, Outdoor Functions, etc. For Electronic Organ, Guitar, String Bass, etc. Gram, Pick-up etc. for mixing purposes, 200-250v. 50 c/s A.C. mains. For 3 and 15 ohm speakers. Twin-handled perforated cover \$1.90. 12 MONTHSF GUARANTEE. Carr. 659

TERMS: Deposit \$4 and 9 monthly payments of \$2.10 (Total \$22.90). Send 8.A.E. for leafet.

LINEAR T40/60D 50W GROUP/DISCO AMPLIFIER

R.S.C. A10 30 WATT ULTRA LINEAR

TYPE C4100 IS ALSO SUITABLE FOR BASS GUITAR OR ELECTRONIC ORGAN

All types 15 Ohms covered in Regine and Vynsir

Units listed below
(a) 100w POWER AMPLIFIER
(b) PAIR OF HI-FI HEAD-PHONES

(c) MATCHING DYNAMIC
'MIKE' (attached to headphones)
(d) PAIR 50 WATT SPEAKERS
Black Rexine covered

Size approx. 18" × 8" (a)(b)(c)&(d) £72.85

n Amps. Speakers and Head-like. Deposit £15 and 18 payments of £4.00 (Total Terms on An phone/Mike.



TDI DISCO CONSOLE

Incorporating twin Garrard SP25
Mk.III turntables and Sonotone or
Acos.Cartridges with diamond stylii.
Separate Vol. controls for each
turntable. Also MONITORING beparate vol. controls for each turntable. Also MONITORING FACILITIES, plus Treble and Bass Controls. Separate input for 'mike' with vol. control switch. Black Rexine covered Cabinet with lid.

See illus. on left
Carr. £1-25
Or Dep. £18-25 and 9 mthly pymis £6-88
(Total £75-17) or Dep. £15 and 18 mthly payments £3.62 (Total £80.62).

FANE ULTRA HIGH POWER LOUDSPEAKERS TWIN TURNTABLE WITH PRE-AMP. All power ratings are R.M.S. continuous. 2 YEARS' GUARANTEE High flux ceramic magnets. ALL CARRIAGE FREE.

AND PUBLIC ADDRESS

'POP' 100 18" 100 Watt 14,000 gauss 8/15Ω

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15" 60 Watt 14,000 gauss 8/15Ω £12.90 Dep. £3.30 and 9 monthly payments. £1.30 (Total £15).

'POP' 60

'POP' 50 12" 50 Watt 13,000 gauss 15 Ω £10.90

Dep 22 and 9 monthly pay-ments 21:20 (Total 212:80) AIR SUITABLE ALL PURPOSES

FOR BASS GUITAR, ELECT, ORGAN, ETC. FANE SPEAKERS 'POP' 25/2 12 in. 25 WATT

Dual Cone 15 Ω (for uses other than Bass Guitar or Electronic Organ). Carr. free

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PAIR L12/25 L/S in cabinets
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payments of £6-56 (Total £67-69)

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PACKAGE PRICE £45 CAFT. £48-45

PACKAGE PRICE £34.95 £21.80 £52 CAFT. £56.75

PACKAGE PRICE £34-95 £56 carr. #22.90 £58-85

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PACKAGE PRICE £99.95 CAFE.

HIGH QUALITY LOUDSPEAKER UNITS

ALL TWO TONE REXINE AND VYNAIR FINISH
L125 50 WATT Fitted pair of 12" 50
watt high flux speakers for conservative
rating. Impedance 8-15 ohms. Carr. £1:50
0r deposit \$4:50 and 9 monthly payts. of £3:70. Total £37:80
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10,000 lines 15 ohms. Carr. 50p.

£11 95 | 10,000 lines 3 or 15 ohms. State impedance required. Carr 40p

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INTEREST CHARGES REFUNDED ON CREDIT SALES Settled in 3 months

£39.75 £23.90

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* FULL MONITORING FACILITIES

★ Individual Bass and Treble controls

★ Frequency response 30-20,000 Hz.

FRONT PANEL (Jack Inputs)

Twin jack output sockets for speakers

★ Input sockets for medium or high impedance Microphones or Guitar P.Us and Crystal of Ceramic Gram. P.Us

Output overload protection plus double fusion

Gram, Gram, Microphone/Gultar P.U; plus monitoring socket.

CONTROLS Vol. (1) (2) (3) and (4). Bass, Treble, Master Vol., Monitor Selection Switch Gram (1) or (2). Mains Switch

MICROPHONE

FAL PHASE 50 Mk.III AMPLIFIER 50 WATT.

Terms Deposit £7.50 and 9 monthly payments £5.37 (Total £55.83)

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£40.20

Simulated black rexine, plastic bonded metal case

Soid state. 4 Separately controlled inputs Plus master vol. control. Ind. Bass and Treble Controls. Protective circuit to guard against damage from accidental shorts. Output for Speaker/s 3 to 30 ohms. Size 17" × 7" × 7" × 71" 200-250v. A.C. mains. Output 50 watts music rating. Or deposit 87:25 & 9 monthly payments £3.55. Total £39-20.

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For Guitar, Vocal or Instrumental Group

A 4 input, 2 vol. control Hi-Fi 30 watt unit with Separate Bass and Treble controls. Current valves. Peak output rating. Strong Regine covered cabinet with handles. Attrac-

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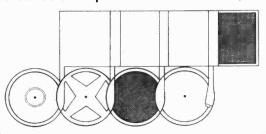
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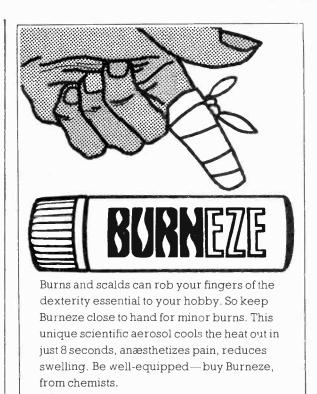
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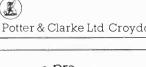
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Use Antex low-leakage soldering irons

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The 15 watt miniature model CCN. also has negligible leakage. Test voltage 4000v. A.C. Totally enclosed element in ceramic shaft. Fitted long-life iron-coated bit 3/32"

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Mallory Cells in U2, U11 and Penlite U7 sizes, Kestrel Battery Charging Unit



Tape recorders, Portable radio and TV, Radio controlled model aircraft and boats, Cine cameras, Flashguns, Cordless shavers and other battery appliances.

Whenever you must have utterly dependable battery power—then these new Alkaline-Manganese rechargeable cells will provide it. The cells can be recharged many times, a simple job with the Kestrel Charging Unit which has been specially designed for these new type batteries.

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275 WEST END LANE LONDON, N.W.6



Solve your communica-tion problems with this 4-Station Transistor Intercom system (1 master and 8 Subs), in de-luxe piastic cabinets for deek or wall mounting. Califtaik/listen from Master to Subs and Subs to Master. Ideally suitable for Business, Sur-gery, Schools, Hospitai, Office and Home. Operates on one 9V battery. On/off switch. Volume control. Complete witu 3 connecting wires each 66ft. and other accessories, P. & P. 20.40.

MAINS INTERCOM

No batteries—no wires. Just plug in the mains for instant two-way, loud and clear communication. On off switch and volume control. Price £12-40



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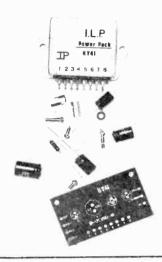
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The HY41 supersedes the popular HY40 introduced by ILP last year. This highly improved module achieves true High Fidelity with a dramatic reduction in distortion (typically 0.05% at 1KHz into 8 ohms!) and is electronically and mechanically compatible with the HY40.

With this important improvement the HY41 retains all of the quality characteristics found in the earlier version and P.C. board, Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete power amplifier of extremely high performance sufficiently versatile to provide power not merely for Hi-Fi but also for public address systems and industry.

The free manual gives a full circuit diagram of the HY41 and its various applications including a complete stereo amplifier

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over recent years.

OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts

R.M.S. continuous.

LOAD IMPEDANCE: 4–16 ohms.

INPUT IMPEDANCE: 30K ohms at 1KHz.

VOLTAGE GAIN: 30db at 1KHz

TOTAL HARMONIC DISTORTION: less than 0.15% (typical 0.05%)

FREQUENCY RESPONSE: 5Hz-50KHz + 1db

SUPPLY VOLTAGE: + 22.5volts D.C. SUPPLY CURRENT: 0.8 amps maximum

PHICE: inc. comprehensive manual, P.C. board, five extra components and P. & P. ... MONO: £4.90 STEREO: £9.80

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Internally the HY5 provides equalization for almost every conceivable input, the desired function is achieved by use of a multi-way switch or by direct interconnection,

Two distinctive features of the HY5 are its inbuilt stabilization circuit, allowing it to be run off any unregulated power supply from 16–25 Volts and a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo pre-amplifier

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely low noise with a high overload capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system.

Magnetic Pick-up (within ± 1 db R1AA curve) 2mV. 47K Ω Tape Replay (external components to suit head). 4mV. 47K Ω

Microphone (flat) 10mV. 47KΩ Ceramic Pick-up (equalized and compen-

satable) 20–2000mV, variable. Tuner (flat) 250mV, 100K Ω Auxiliary 1 250mV, 47K Ω Auxiliary 2 2–20mV, 100K Ω

OUTPUTS

Main Pre-amp output 500mV Direct tape output 120mV

ACTIVE TONE CONTROLS (8exendal!) Treble ± 12db. Treble ± 12d Bass + 12db.

Bass + 1240.
INTERNAL STABILIZATION
Enables the HY5 to share an unregulated supply with the Power Amplifier.

SUPPLY VOLTAGE

16-25 volts PRICE:

MONO: £3.60

SUPPLY CURRENT 6mA approx OVERLOAD CAPABILITY better than 26db on most sensitive input infinite on tuner and auxl.

IPHY5

OUTPUT NOISE VOLTAGE: 0.5mV



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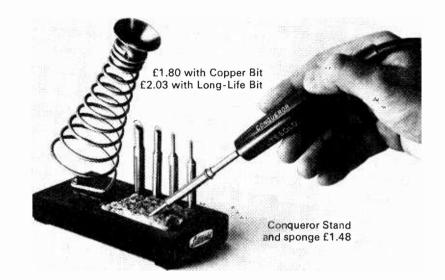
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								PL83	-81	UL84	-87
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								PL500	-59	UY85	-22
30F5	-80	EBC33	-88	EF89	-24	PCC84	.27	PL504	-59	W77	-42
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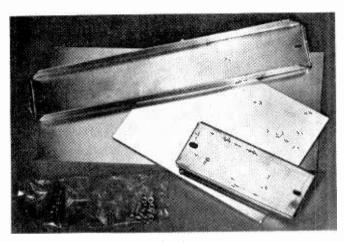
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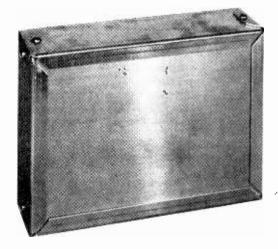
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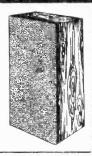
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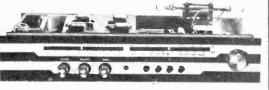
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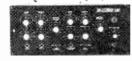
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The PR40 R.F. Preselector is the solid state version of the world famous PR30 which it now supercedes. It employs Silicon "NI" Channel FET (Field Effect Transistor) front end, followed by silicon NPN Broad Band R.F. Amp., and will substantially improve receiver performance over the range 1-5 to 35 MHz, providing a considerable increase in gain up to an overall average of 30 dB, with improved image rejection and noise ratio.

Supplied complete with co-ax plug (less standard 9 volt PP6 Battery), 12 months Guarantee



MULTIBAND—6 SOLID STATE

SHORT WAVE RECEIVER KIT

All transistor T.R.F. Receiver tunes \$50 KHz to 30 MHz (540 to 10 metres) complete coverage—no gaps. Medium waves—Trawlers—Ship/Shore Telephone—All Six Amateur Bands 160–10 metres—International Broadcast from Australia, Far East, Russia, USA etc. using 4 miniature plug in Coils. Hi-Gain FET Regen. Det./AF/AF Module giving full loudspeaker output to any external 2/3 ohm speaker, Receives AM/CW/SSB.
Separate Electrical Bandspread, Calibrated Main Tuning.
A Quality CODAR-KIT with 12 months Guarantee. No technical knowledge required, simple to build, printed circuit and Pictorial Instruction Manual, fully detailed step by step.
Complete Kit with 4 Coils (less standard PP6 battery). £13-20 Carriage 35p. Stamp brings illutrated leaflets.





COMMUNICATION RECEIVER

The CODAR CR70A is an outstanding general coverage communication

receiver, ideal for the keen S.W.L.

It tunes from 540 metres medium through to 10 metres with no gaps. It tunes from 540 metres medium through to 10 metres with no gaps. Covers shipping, coatsguard and distress frequencies, all six amateur bands 160-10 metres, International broadcast, Met. stations etc. etc. giving world-wide reception. Exclusive features include Air-spaced CODAR-COIL Hi-"Q" Aerial input, illuminated Meter and Slide Rule Scale, Two Speed vernier tuning, Switched B.F.O. for CW/SSB signals. Separate output for Tape recorder. Ready to plug in to 200/240 volts A.C. it only needs your aerial and a 2/3 ohm Joudspeaker to have the world to your fineer tips. 12 months

2/3 ohm loudspeaker to bring the world to your finger tips. 12 months full guarantee.

Complete ready built £27 50, Carriage 70p. Ask the chap who's got one, there's lots of them!

NOTE: These CODAR Receivers meet the B.R.E.M.A. specification for Communication Receivers and are FREE of Purchase Tax. Some portable receivers being advertised as Communication Receivers do not meet these requirements.



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G3SZM G31RE G8BBI

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THIS IS THE FIRST PAGE OF THE GREAT

BRAND NEW FULLY GUARANTEED DEVICES

AC107 AC113 AC115 AC117K AC122 AC125 AC125 AC127 AC128 AC132 AC132 AC132 AC131 AC141 AC141 K AC141 K	0 20 0 20 0 23 0 20 0 12 0 17 0 17 0 14 0 14 0 14 0 17 0 14	AD162 AD161 & AD162 (R AD140 AF114 AF115 AF116 AF117 AF118 AF124 AF125 AF126 AF127 AF127 AF139	0 38 4P) 0 55 0 50 0 24 0 24 0 35 0 30 0 28 0 30	BC148 BC149 BC150 BC151 BC152 BC153 BC154 BC157 BC158 BC169 BC160 BC161 BC167 BC168 BC169	0 10 0 12 0 18 0 20 0 17 0 28 0 30 0 18 0 12 0 45 0 50 0 12 0 12 0 12	BD137 BD138 BD139 BD140 BD155 BD175 BD176 BD177 BD178 BD179 BD180 BD185 BD186 BD187 BD188	0 45 0 50 0 55 0 60 0 80 0 60 0 65 0 70 0 70 0 65 0 65 0 70	BF188 BF194 BF196 BF196 BF197 BF200 BF222 BF257 BF258 BF259 BF262 BF263 BF270 BF271 BF272	0·40 0·12 0·12 0·14 0·14 0·45 0·95 0·85 0·85 0·55 0·35 0·80	OC19 OC20 OC22 OC23 OC24 OC25 OC26 OC28 OC29 OC35 OC36 OC41 OC41 OC42 OC42	0 35 0 63 0 88 0 42 0 50 0 38 0 25 0 50 0 42 0 50 0 20 0 20 0 15 0 12	2G371B 2G371B 2G3773 2G374 2G377 2G378 2G381 2G381 2G401 2G417 2G417 2N388 2N388 A 2N404 2N404 A	0 16 0 12 0 17 0 17 0 30 0 16 0 30 0 30 0 25 0 35 0 28	2N2219 2N2220 2N2221 2N2222 2N2369 2N2369 2N2411 2N2412 2N2646 2N2711 2N2712 2N2714 2N2714 2N2904 2N2904	0 · 24 0 · 24 0 · 47 0 · 21 0 · 21 0 · 21 0 · 21	2N3054 2N3055 2N3391 2N3391A 2N3393 2N3393 2N3394 2N3395 2N3402 2N3403 2N3404 2N3405 2N3405 2N3416 2N3416 2N3416 2N3416	0 46 0 50 0 14 0 14 0 14 0 17 0 21 0 28 0 42 0 15 0 28	2N4059 2N4060 2N4061 2N4062 2N4284 2N4285 2N4286 2N4286 2N4288 2N4289 2N4299 2N4291 2N4292 2N4292 2N4292 2N4292	0 10 0 12 0 12 0 17 0 17 0 17 0 17 0 17 0 17 0 17 0 17
ACIUL	0·17 0·15	AF178 AF179	0.50	BC170 BC171	0 12	BD189 BD190	0.75	BF273 BF274	0.35	0C70 0C71	0·10 0·10	2N524 2N527	0 · 42 0 · 49 0 · 42	2N 2905 2N 2905 A 2N 2906		2N3525 2N3646	0 · 75 0 · 09	2N5458 2N5459	0.32
AC154 AC155	0.20	AF180 AF181	0.50	BC172 BC173	0:14	BD195 BD196	0·85 0·85	BFW10 BFX29	0·60 0·27	OC72 OC74	0·14 0·14	2N598 2N599	0.45	2N2906A	0.18	2N3702	0.10	28301	0.50
AC156	0.20	AF186	0.45	BC174	0 - 14	BD197	0.90	BFX84 BFX85	0.22	OC75 OC76	0·15 0·15	2N696 2N697	0·12 0·13	2N2907 2N2907A	0.20	2N3703 2N3704	0·10 0·11	28302A 28302	0·42 0·42
AC157 AC165	0 24	AF239 AL102	0·37 0·65	BC175 BC177	0.22	BD198 BD199	0.95	BFX86	0.22	OC77	0.25	2N698	0.24	2N2925	0.14	2N3705	0.10	28303	0.55
AC166	0.20	AL103	0.65	BC178	0.19	BD200	0.95	BFX87 BFX88	0 24	OC81 OC81D	0·15 0·15	2N699 2N706	0.35	2N2924 2N2925	0·14 0·14	2N3706 2N3707	0·09 0·11	28304 28305	0·70 0·84
AC167 AC168	0 20	ASY26 ASY27	0 - 25	BC179 BC180	0 19	BD205 BD206	0.80	BFY50	0.20	OC82	0.15	2N706A	0.09	2N2926 (G)	2N3708	0.07	28306	0.84
AC169	0.14	AS Y 28	0 - 25	BC181	0.24	BD207	0.95	BFY51	0.20	OC82D	0.15	2N708	0·12 0·30	2N2926 (0.12	2N3709 2N3710	0 09	28307 28321	0.84
AC176 AC177	0 · 20 0 · 24	ASY 29 ASY 50	0 · 25 0 · 25	BC182 BC182L	0·10 0·10	BD208 BDY20	0·95 1·00	BFY52 BFY53	0 20	OC83 OC84	0 20	2N711 2N717	0.35	2142920 (0·11	2N3711	0.09	28322	0 42
AC178	0.28	A8Y51	0.25	BC 83	0.10	BF115	0.24	BPX25	0.85	OC139	0.20	2N718	0.24	2N2926 (O) 0·10	2N3819	0.28	28322A 28328	0 42
AC179 AC180	0.28	ASY52 ASY54	0.25	BC183L BC184	0·10 0·12	BF117 BF118	0 · 45 0 · 70	BSX 19 BSX 20	0·15 0·15	OC140 OC169	0 · 20 0 · 25	2N718A 2N726	0·50 0·28	2N2926 (2N3820 2N3821	0.85	2832	0.70
AC180 K	0.20	ASY55	0.25	BC184L	0 12	BF119	0.70	BSY25	0.15	OC170	0.25	2N727	0.28		0.10	2N3823	0.28	28325	0.70
AC181	0.17	ASY56	0.25	BC186	0·28 0·28	BF121 BF123	0 · 45 0 · 50	BSY26 BSY27	0·15 0·15	OC171 OC200	0 · 25 0 · 25	2N743 2N744	0.20	2N2926 (B) 0·10	2N3903 2N3904	0 · 28 0≈30	28326 28327	0.70
AC181 K AC187	0·20 0·22	ASY57 ASY58	0.25	BC187 BC207	0.11	BF125	0.45	BSY28	0.15	OC201	0.28	2N914	0.14	2N3010	0.70	2N3905	0.28	28701	0.42
AC187K	0.20	ASZ21	0.40	BC208	0.11	BF127 BF152	0.50	BSY29 BSY38	0 15	OC202 OC203	0 · 28 0 · 25	2N918 2N929	0·30 0·21	2N 3011 2N 3053	0·14 0·17	2N3906 2N4058	0·27 0·12	40361 40362	0·40 0·45
AC188 AC188K	0 22	BC107 BC108	0.09	BC209 BC212L	0·12 0·11	BF153	0.45	BSY39	0.18	OC204	0.25	2N930	0.21	2110000	0 1.	211 1000	0 10		
ACY17	0.25	BC109	0.10	BC213L	0.11	BF154	0 45	B8Y40 B8Y41	0·28 0·28	OC205 OC309	0-35	2N1131 2N1132	0 · 20 0 · 22						
ACY18 ACY19	0.20	BC113 BC114	0·10 0·15	BC214L BC225	0 14	BF155 BF156	0.70	BSY95	0.12	P346A	0.20	2N1302	0.14			ES AND			
ACY20	0.20	BC115	0.15	BC226	0.35	BF157	0.55	BSY95 A	0.12	P397 OCP71	0.42	2N1303 2N1304	0·14 0·17	AA119 AA120	0 08	BY 133 BY 164	0.21	OA10 OA47	0.35
ACY21	0 - 20	BC116 BC117	0·15 0·15	BCY30 BCY31	0.24	BF158 BF159	0.55	Bu105 C111E	2.00	ORP12	0.43	2N1304	0.17	AA129	0.08	BYX 38/3	30	OA70	0.07
ACY27	0.18	BC118	0.10	BCY32	0.80	BF160	0.40	C400	0.30	ORP60	0 40	2N1306 2N1307	0·21 0·21	AAY30 AAZ13	0.09	BYZ10	0 - 42	OA79 OA81	0 - 07
ACY28 ACY29	0 19	BC119 BC120	0 30	BCY33 BCY34	0 · 22 0 · 25	BF162 BF163	0-40	C407 C424	0 25	OR 1'61 ST140	0 · 40 0 · 12	2N1307	0.23	BA100	0.10	BYZ11	0.30	OA85	0.09
ACY30	0-28	BC125	$0 \cdot 12$	BCY70	0.14	BF164	0.40	C425	0.50	ST141	0-17	2N1309	0.23	BA116 BA126	0·21 0·22	BYZ12 BYZ13	0·30 0·25	OA90 OA91	0 · 06 0 · 08
ACY31 ACY34	0.28	BC126 BC132	0·18 0·12	BCY71 BCY72	0·18 0·14	BF165 BF167	0 - 40	C426 C428	0.35	TIS43 UT46	0·30 0·27	2N 1613 2N 1711	0.20	BA148	0.14	BYZ16	0.40	OA95	0.07
ACY35	0 21	BC134	0.18	BCZ10	0.20	BF173	0.22	C441	0.30	2G301	0.09	2N 1889	0·32 0·45	BA154 BA155	0·12 0·14	BYZ17 BYZ18	0.35	OA200 OA202	0.06
ACY40	0 28	BC135 BC136	0·12 0·15	BCZ11 BCZ12	0 · 25 0 · 25	BF176 BF177	0 · 35 0 · 35	C442 C444	0·30 0·35	2G302 2G303	0 19	2N1890 2N1893	0.37	BA156	0.13	BYZ19	0.28	SD10	0.05
ACY41	0.18	BC137	0.15	BD121	0.60	BF178	0.30	C450	0.22	2G304	0.24	2N2147	0.72	BY100 BY101	0·15 0·12	CG62 (Eg) OA9	11	8D19 IN34	0 - 05
ACY44 AD130	0.35	BC139 BC140	0 40	BD123 BD124	0 - 65	BF179 BF180	0.30	MAT100 MAT101	0 - 19	2G306 2G308	0 - 40	2N2148 2N2160	0.60	B ¥ 105	0 17		0.05	IN34A	0.07
AD140	0.48	BC141	0.30	BD131	0.50	BF181	0.30	MAT120	0.19	2G309	0.35	2N2192	0.35	BY114 BY126	0·12 0·14	CG651 (Eq) OA	70.	1N914 1N916	0 · 06 0 · 06
AD142 AD143	0.48	BC142 BC143	0 30	BD132 BD133	0-60 0-65	BF182 BF183	0 · 40 0 · 40	MAT121 MPF102	0 · 20 0 · 42	2G339 2G339A	0 · 20 0 · 16	2N2193 2N2194	0 · 35 0 · 35	BY127	0.15	OA79	0.06	IN414B	0.06
AD149	0.50	BC145	0.45	BD135	0.40	BF184	0.25	MPF104	0.87	2G344	0.18	2N2217 2N2218	0 22	BY128 BY130	0·15 0·16	OA5 OA5SL	0 - 35	IS021 IS951	0.10
AD161	0.33	BC147	0.10	BD136	0.40	BF185	0.30	MPF105	0.97	2G345	0.16	2172218	0.20	21100	3.10	JAOBE	3 24	20001	3 00

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Pack No.	Qty.	Description		Price
C 1	250	Resistors mixed values approx. count by weight		0.50
C 2	200	Capacitors mixed values approx. count by weight		0.50
C 3	50	Precision Resistors 1%, 01% mixed values		0.50
C 4	75	th W Resistors mixed preferred values		0.50
C 5	5	Pieces assorted Ferrite Rods		0.50
C 6	2	Tuning Gangs, MW/LW VHF		0.50
C 7	1	Pack Wire 50 metres assorted colours		0 - 50
C 8	10	Reed Switches		0.50
C 9	3	Micro Switches		0.50
C10	15	Assorted Pots & Pre-Sets		0.50
C11	5	Jack Sockets 3 × 3.5m 2 × Standard Switch Types		0.50
C12	40	Paper Condensers preferred types mixed values		0 · 50
C13	20	Electrolytics Trans. types		0 - 50
C14	1	Pack assorted Hardware-Nuts/Bolts, Grommets etc		0.50
C15	4	Mains Toggle Switches, 2 Amp D/P		0 - 50
C16	20	Assorted Tag Strips & Panels		0.50
C17	10	Assorted Control Knobs		0.50
C18	4	Rotary Wave Change Switches		0.50
C19	3	Relays 6-24V Operating		0.50
C20	4	Sheets Copper Laminate approx. 10" × 7"		0.50
		10_{P} post and packing on all component packs, plus a feas. C1, C2, C19, C20.	irth	er 10p

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Modual Tested and Guaranteed (1972) 1982 63: 10—25 22:28 Price each. Larger quantities quoted on request. Full hook-up diagrams and complete technical data supplied free with each modual or available separately at 10p each.

SYSTEM 12 STEREO

Each Kit contains two Amplifier Modules, 3 watts RMS, two loudspeakers, 15 ohms, the pre-amplifier, trans-former, power supply module, front panel and other accessories, as well as an illustrated stage-by-stage instruc-ONLY tion booklet designed for the beginner.

Further details available on request.



NEW LOW PRICE TESTED S.C.R.'s

PIV 1A 3A 5A 5A 7A 10A 16A 30A TO5 TO66TO66TO64TO48TO48TO48TO48 $\begin{array}{c} \textbf{T05} \quad \textbf{T068T0664T0648T048T048T048}\\ \textbf{50} \quad 0.23 \quad 0.25 \quad 0.36 \quad 0.36 \quad 0.31 \quad 1.5\\ 100 \quad 0.25 \quad 0.38 \quad 0.37 \quad 0.49 \quad 0.69 \quad 0.58 \quad 1.63 \quad 1.40 \\ 200 \quad 0.36 \quad 0.37 \quad 0.49 \quad 0.49 \quad 0.57 \quad 0.61 \quad 0.75 \quad 1.60 \\ 400 \quad 0.43 \quad 0.47 \quad 0.56 \quad 0.56 \quad 0.67 \quad 0.76 \quad 0.98 \quad 1.75 \\ 600 \quad 0.53 \quad 0.57 \quad 0.80 \quad 0.86 \quad 0.77 \quad 0.79 \quad 1.25 \\ 800 \quad 0.63 \quad 0.70 \quad 0.80 \quad 0.80 \quad 0.90 \quad 1.20 \quad 1.50 \quad 4.00 \end{array}$

SIL. RECTS. TESTED

PIV	300m.A	750m A	1.A	1.5 A	-6A	10A	30 A	
50	0.04	0.05	0.05	0.00	0.14	0 - 21	0.60	
100	0.04	0.06	0.0%	$0 \cdot 13$	0.16	0.23	0.75	
200	0.05	0.09	0.00	0.14	0 20	0.24	1.00	
400	0.06	0.13				0.37		
600	0.07	0.16				0.45		
800	0.10	0.17				0.55		
1000	0.11	0.25				0.63		
1200		0.33				0.75		

FULL RANGE ZENER DIODES

2-33V 400mV (DO-7 Case 13p ca. 14W (Top-Hst) 18p ca. 10W (80-10 Stud) 25c ca. All fully tested 5% tol. and marked. State voltage required.

10 amp POTTED **BRIDGE RECTIFIER** on heat sink.

100PIV. 90p each

2 Amp. BRIDGE RECTS. 50 v RMS 32p each 100 v RMS 37p ... 400 v RMS 46p ... 8ize 15 mm × 6 mm.

UNIJUNCTION UT46. Eqvt. 2N2646, Eqvt. TI843. BEN3000 27p each, 25-99 100 UP 20p.

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GENERAL PURPOSE
NPN SILICON SWITCHING TRANS. TO-18
SIM. TO 2N706/8. BSY27/28/95A. All usable
devices no open or short
circuits. ALSO AVAILABLE in PNP Sim. to
2N2906, BCY7O. When
ordering please state
preference NPN or PNP.

For For For

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SIL. G.P. DIODES &p

VOLTAGE

NEW LINE Plastic Encapsulated OF

RANGE

TRIACS

VBOM 2A 6A 10A TO-1 TO-66 TO-88 £p ₫p £p 80 100 50 76 200 60 90 400 70 75 1-10

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Fa.k	No.	EQVT
T1	8 203713	OC71
T2	8 D1374	OC75
T3	8 D1216	OC81D
T4	8 2G381T	OC81
7.2	8 2G382T	OC82
T6	8 2G344B	OC44
T.7	8 2G345B	OC45
T8	8 2G378	OC78
Т9	8 2G399A	2N1302
T10	8 20417	AF117

All 50p each pak 2N2060 NPN SIL. DUAL TRANS. CODE D1699 TEXAS. Our price 25p

120 VCB MIXIE DRIVER TRANSISTOR. Sim. ISX21 & C407, 2N1893 FULLY TESTED AND CODED ND 120. 1-24 17p each. TO.5 N.P.N. 25 up 15p each.

l. trans. suitable for E. Organ. Metai TO-18 pt. ZTX300 5p each. Eqvt. ZI

POWER **TRANS**

300mW 30..0-50 40PIV(Min.) 100..1-50 Sub-Min. 500..5-00 Full Tested 1,000..9-00 Ideal for Organ Builders. BONANZA!

1000

GENERAL PURPOSE GERM. PNP
Coded GP100. BRAND NEW TO-3 CASE. FOSS.
REPLACE:—OC25-28-29-30-35-36. NKT 401-403404-405-406-430-461-462-463. T13027-3028. 2N250A,
2N456A-457A-458A, 2N511 A & B. 20220-222, ETC.
VCBO 80V VCEO 50V IC 10A PT. 30 WATTS He
30-170.
PRICE 1-24 25-00

40p

43p each BILICON High Voltage 250V NPN
TO-3 case. G.P. Switching & Amplifier
Applications. Brand new Coded R 2400
VCBO 250V/VCEO 100/1C 6A/30 Watts.
HFE type 20/IT 5MHZ.
OUR PRICE EACH:
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509 455

100 up 50p 45p

AD161/162 PNP M/P COMP GERM TRANS. OUR LOWEST PRICE OF 55p PER PAIR 40p each 36p each

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18.00

2N3055

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Pak F	ſo.	Description	Price
U 1		Glass Sub-Min. General Purpose Germanium Diodes	0.50
U 2	60	Mixed Germanium Transistors AF/RF	0.50
U 3	75	Germanium Gold Bonded Sub-Min. like OA5, OA47	0.50
U 4	40	Germanium Transistors like OC81, AC128	0 - 50
U 5	60	200m A Sub-Min. Silicon Diodes	0.50
U 6		Sil. Pianar Trans. NPN like BSY95A. 2N706	0.50
U 7		SII. Rectifiers TOP-HAT 750mA VLTG. RANGE up to 1000	0-50
U 8		Sil. Planar Diodes DO-7 Glass 250mA like OA200/202	0.50
U 9	20	Mixed Voltages, 1 Watt Zener Diodes	0.50
U10		BAY50 charge storage Diodes DO-7 Glass	0 - 50
U11	25	PNP Sil. Planar Trans. TO-5 like 2N1132, 2N2904	0.50
Ū12		Silicon Rectifiers Epoxy 500mA up to 800 PIV	0.50
U13		PNP-NPN Sil. Translators OC200 & 28 104	0.50
Ū14		Mixed Silicon and Germanium Diodes	0.50
U15		NPN Sil. Planar Trans. TO-5 like BFY51, 2N697	0.50
U16		3 Amp Silicon Rectifiers Stud Type up to 1000PIV	0.50
U17		Germanium PNP AF Transistors TO-5 like ACY 17-22	0.50
U18		6 Amp Silicon Rectifiers BYZ13 Type up to 600 PIV	0.50
U19		Silicon NPN Transistors like BC108	0.50
U20		1.5 Amp Silicon Rectifiers Top Hat up to 1000 PIV	0.50
U21		AF. Germanium Alloy Transistors 2G300 Series & OC71	0.50
U23		MADT's like MHz Series PNP Translators	0.50
U24		Germanium 1 Amp Rectifiers GJM Series up to 300 PIV	0.50
U25		300 MHz NPN Silicon Transistors 2N708, BSY27	0.50
U26		Fast Switching Silicon Diodes like IN914 Micro-Min.	0.50
U27		NPN Germanium AF Transistors TO-1 like AC127	0.50
U29		1 Amp 8CR's TO-5 can, up to 600 PIV CR81/25-600	0.50
U30		Plastic Silicon Planar Trans. NPN 2N2926	0.50
U31		Silicon Planar Plastic NPN Trans. Low Noise Amp 2N3707	0.50
U32		Zener Diodes 400mW DO-7 case 3-18 volts mixed	0.50
U33		Plastic Case 1 Amp Silicon Rectifiers IN4000 Series	0.50
U34		Silicon PNP Alloy Trans. TO-5 BCY26 28302/4	0.50
U35		Silicon Planar Transistors PNP TO-18 2N2906	0.50
U36		Silicon Planar NPN Transistors TO-5 BFY50/51/52	0.50
U37		Silicon Alloy Transistors 80-2 PNP OC200, 28322	0.50
U38		Fast Switching Silicon Trans. NPN 400 MHz 2N3011	0.50
U39		RF. Germ. PNP Transistors 2N1303/5 TO-5	0.50
U40		Dual Transistors 6 lead TO-5 2N2060	0.50
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U42		VHF Germanium PNP Transistors TO-1 NKT667, AF117	0.50
U43		Sil. Trans. Plastic TO-18 A.F. BC113/114	0.50
U44		Sil. Trans. Plastic TO-5 BC115/NPN	0 - 50
	20	Distriction of the control of the co	
U45	7	3A SCR. TO66 up to 600PIV	1.00

Code No's, mentioned above are given as a guide to the type of device in the pak. The devices themselves are normally unmarked.

SILICON PHOTO TRAN-BISTOR. TO-18 Lens end N I'N Sim. to BP × 25 and P21. BRAND NEW. Full data available. Fully guaranteed. Qty. 1-24 25-99 100 up Price each 45p 40p 35p

F.E.T.'S

2N3819	85p	2N5458	50p			
2N3820	50p	2N5459	40p			
2N3821	85p	BFW10	40p			
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BN7430	0.15	0.14	0.12	SN74104	0.97	0.94	0.38
8N7432	0.45	0.42	0.40	8N74105	0.97	0.94	0.88
SN7433	0.80	0.75	0.70	8N74107	0.40	0.38	0.36
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BP951	65p	60p	65p
BP962	12p	11p	10p
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Tube Height (mm)	47	32	22	data for all types available
Tube Diameter (mm)	19	13	12 wide	on request.
I.C. Driver Rec.	BP41 or 141	BP41 or 141	BP47	
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SPECIFICATION Frequency Response Harmonic Distortion Inputs: 1. Tape Head 2. Radio, Tuner

3. Magnetic P.U.

Bass Control Treble Control

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All input voltages are for an output of 250mV, Tape and P.U. inputs equalised to RIAA curve within ± 1dB, from 20Hz to 20KHz. ± 15dB at 20Hz ± 15dB at 20KHz 100Hz

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RACTICAL WIRELESS

VOL 48 NO 7

Issue 789

NOVEMBER 1972

Is Ignorance Bliss?

M ORE than £100,000,000's-worth of audio equipment, half of which is hi-fi, is being sold in the UK every year. But according to the findings of the commercial research department of Philips Electrical, who published details of a survey earlier this year, this buoyant industry thrives in a sea of appalling ignorance.

The report doesn't mince words. It says that "the level of ignorance encountered in a number of shops was startling". Many of them had inadequate audio facilities and expertise. When questioned, most dealers considered that the most important factor was the price, this being considered far

ahead of design and performance.

This will come as no surprise to knowledgeable types who have had to buy goods from such shops. But the report goes on to say that the customers' largest single source of information is not from friends, magazines, sales literature or advertisements, but from dealers. It goes without saving, then, that in this classic example of the blind leading the blind, the average customer comes out of the ordeal in a state of mental fog.

To quote from the results of random samples: 44% did not know whether their cartridge was magnetic or ceramic; 75% did not know the frequency response of their amplifier; 60% could not give the audio output, and of those who could, 33% did not know if the figure was r.m.s. or music rating. Of more pertinent interest is that many people do not know whether or not they are listening to stereo. Many thought they were hearing stereo on Radios 1 and 2 and 75% of intending buyers interviewed thought they heard

stereo on Radio 1!

It does not need a research team to expose the sad lack of competence (and, often, interest) in many shops, and it follows logically that many customers must be similarly ill-informed. However, we feel that there is a dividing line between the depth of knowledge a customer needs to use his equipment properly and the knowledge which is simply a technical luxury. In view of this, we were delighted to read that, despite the epidemic of ignorance, 64% of customers were "very satisfied" with their equipment.

One trade journal has suggested that we want a "big effort to create more unsatisfied customers". However, although we want people to appreciate good sound reproduction, is it ethical to unsettle people who are perfectly content with their equipment in order to gain a few more kHz and a little less distortion which they probably could not distinguish? In general we are against ignorance but sometimes, especially in audio, it can be bliss!

W. N. STEVENS-Editor.

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The ABC of Hi-Fi by Gordon J. King

THE DECEMBER ISSUE WILL BE **PUBLISHED ON NOVEMBER 3rd**

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Mini calculator

Sinclair Radionics Ltd., annouce the world's smallest electronic calculator, just over ¹4in. thick,

The Sinclair "Executive" performs all the functions of large desk machines although it is only 2 inches wide by 5½ inches long. The total weight, including batteries, is 2½ ounces. The illuminated display has a capacity of 8 digits and the machine will add, subtract, divide and multiply virtually instantaneously. Other features include automatic squaring, reciprocals, fixed or floating decimal point operation, and a memory for locking in instructions to repeatedly multiply or divide by a pre-determined factor.

This calculator uses a single MOS chip containing 7,000 transistors. Normally this type of I.C. consumes in the region of 350mW which alone would rule out a really slim pocket calculator because of the large batteries that would be required. The Sinclair calculator, however, includes a special circuit (patents applied for) which reduces the consumption to an average of only about 20mW. This is achieved in two ways; firstly by switching off the I.C. completely in between clock pulses, relying on the capacitance associated with the active elements to store the information, and secondly by reducing the clock rate considerably (and hence the "on" period) when the calculator is not actually performing a calculation.

Another factor contributing to the low overall power of the calculator is the way in which the display is driven. The battery voltage is chosen so that it is just sufficient to drive the monolithic gallium arsenide phosphide display directly so that no power wasted in large dropper resistors. Indeed no such resistors are used at all as the display is driven directly from the chip. The batteries themselves are three tiny mercury cells of the type commonly used in hearing aids. These provide 20 hours of

The keyboard, another item which contributes to the thick-



ness of competitive machines, was specially developed for this calculator and is extremely thin, being only 4mm in depth. Price is £79 and every unit is guaranteed five years. Sinclair Radionics Limited, London Road, St. Ives, Huntingdonshire PE 17 4HJ. St. Ives (04806) 4311.

Texan Reprints

Reprints of the "Texan" 20+20W Stereo Amplifier are now available

They may be obtained by sending 35p (30p + 5p postage/packing) to Practical Wireless Editorial, Fleetway House, Farringdon Street, London, EC4A 4AD. — Hurry the demand will be great!

Sheffield exhibition

Sheffield Education Committee will be holding an exhibition next month with the theme "Communications Today" at the Yewlands Comprehensive School, Creswick Lane, Grenoside, Sheffield, S30 3NN.

The dates of the exhibition are Friday and Saturday 3rd and 4th November and further information may be obtained from J. M. Nicholson, M.A., at the Yewlands School, telephone Ecclesfield 3778 (from London the code is 074 15 3778).

Record care

New from Bib is the Record Care Kit Ref. 43. Contained in a presentation 3-colour printed see-through box will be a new all-chrome finish Groov-Kleen Automatic Record Cleaner incorporating counterweight, arm-rest and all the features of the other two Bib Groov-Kleen's Models 42 and 50, turntable spirit level, plastic record dust-off, stylus & turntable cleaning liquid, right angle cleaning brush suitable for stylus and Groov-Kleen pad, and a turntable cleaning cloth. Recommended retail price for the Record Care Kit Ref. 43 is £2.49. including P.T.



NEWS.

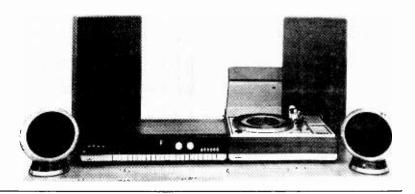
Siemens equipment

The first Siemens record player to become available in the UK is announced by Interconti Electronics, who also announce the introduction of the RS303 tuner amplifier.

The RW346 stereo record player, comes complete with plinth and a protective dust cover. The arm is fitted with Shure magnetic pickup system and diamond stylus. The cabinet is available either in white or a light satinwood finish.

The recommended retail price of the unit is £100.66p.

The RS303 tuner amplifier to which the RW346 can be linked, is a stereo drive unit providing quadrosound reproduction. The receiver is equipped with built-in antennae covering all wave ranges. Station selection in the f.m. range is accomplished easily by using the five pre-set tuners each of which is tuned by a given control button. Price is £180. Interconti Electronics Ltd., Albany House, Petty France, London, S.W.1.



R.F. signal generator

Taylor **Electrical Instruments** have introduced the re-styled 68A/M Mk 11 signal generator which covers 100kHz to 240MHz in eight switched ranges with 30% internal amplitude modulation between 50Hz and 5kHz. All controls and sockets are panel mounted and include Set r.f. Level (with meter indication), Set a.f. Level, r.f. Output (switched 1-10), Attenuator (0 to -80dB.).

1209D. In addition to the many

features of the well established Instruments Limited Transistor Stroboflash, the new stroboscope, Type design provides a high sensitivity input (minimum external trigger: 200mV r.m.s.) which enables the be Stroboflash to triggered

> directly from such sources as vibration meters, vibration analysers or from an electromagnetic pickup.

The Stroboflash produces a high intensity white light and is capable of directly measuring the speed of any shaft or rotating part within the range 300 rev/min to 18,000 rev/min. The very short duration of the light flashes makes it possible to measure speeds up to 180,000 rev/min!

Dawe Instruments Ltd., Western Avenue, London, W3 OSD.

Stubec Radio

The above company have published a new catalogue which contains details and prices of radio components and special details of do-it-vourself loudspeaker components and grill fabrics.

For your copy of this catalogue, send a stamped addressed foolscap envelope to: Stubec Audio, 2 St. Faith's Street, Maidstone, Kent.



The Super Sovereign RP75 covers medium, long, two short and f.m. bands and it's the latest in the Hacker range of portable radios, taking over from the well-known Hacker "Sovereign." Retail price is £69.

Strobe unit

Dawe announce



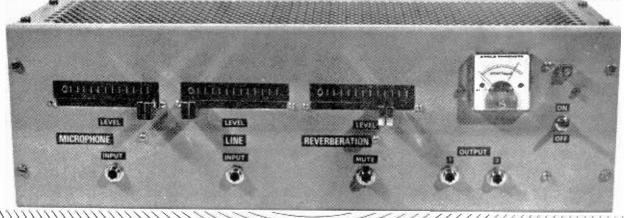
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Practical Wireless will be on Stand 18 (ground floor) at the International Audio Festival and Fair, Olympia from October 23rd-28th.

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Amongst other items featured, there will be a Reverberation Unit, Loud-Hailer, Buzz-Bar, Car Radio, Miniature Organ, Radio Transmitter, Receiver, Power Supply and Audio Test Instruments.

Reverberation Unit



F.C.JUDD

PART ONE

EVERBERATION is an acoustical effect caused by the reflection of sound from hard surfaces within an enclosed space such as a large room or hall containing little sound absorbing material such as carpeting and padded furniture etc. The reflections become multiplied as the sound bounces from surface to surface and reaches the ears as a series of echoes of very short duration. The overall effect however, is a sustained sound because the reflections may take a considerable time to die away completely, depending on the volume of the enclosed space. In a very large hall the dying away, or decay time, may be as much as several seconds. Reverberation is quite unlike the more distinct single echoes caused by sound reflection in the open air from a single surface, such as a large building, or the echo effect produced by feeding a recorded signal from tape back into the recording amplifier.

Spring Line Reverberation

Reverberation is quite easy to produce artificially. All that is required is a) a means of obtaining a series of echoes of short duration, of a few milliseconds in fact, and b) a means of sustaining repetition of the echoes for at least two seconds. A length

of coiled springy wire will do this if it is mechanically excited at one end; this can be done by using a simple electromagnetic sound transducer. The speed of sound vibration up and down the spring line is very slow when compared with the speed of elec-

Specification

Microphone Sensitivity

Line Input Sensitivity
Microphone Input

Impedance
Line Input Impedance
Output Impedance
Frequency Response
(without reverb)
Signal to Noise
(without reverb)
Distortion Factor
(without reverb)
Reverb Delay Time

0.5mV for 800mV output (without reverb) 200mV for 800mV output (without reverb)

Nominal 10kΩ Nominal 5kΩ 20Hz to 20kHz, either input

Nominal 100kΩ

Microphone: —60dB Line: —80dB Less than 0·5 per cent

Approx 2 seconds maximum

trical sound through wires. The spring line offers little resistance to vibration travelling along it and because of this the vibrations will continue to travel up and down the line for some time. The strength of the vibrations do, however, decrease slowly and eventually cease altogether.

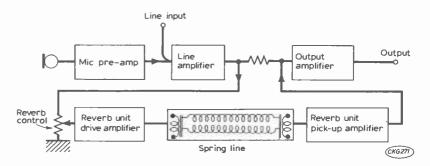


Fig. 1: Block diagram of the spring line reverberation unit.

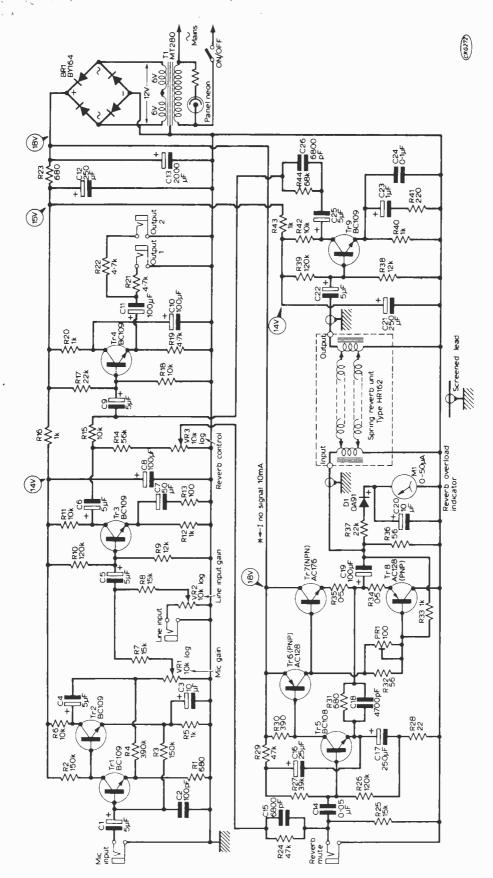


Fig. 2: Circuit diagram of the complete reverberation unit.

The complete reverberation unit

The block diagram in Fig. 1 will help to clarify the principle of a spring line reverb system. First the direct signals from a microphone or other external signal source are taken into the appropriate input. These are amplified and taken both to the output amplifier and to the spring line driving amplifier. The output from the latter drives the spring line and the reverberated signals are taken from the other end of the line, amplified and mixed with the direct signal. The output signal is therefore a mixture of both direct and delay signals. A control is inserted so that the level of signal from the reverberation circuit can be adjusted to provide different reverberation levels or it can be cut off altogether. One can simulate the short reverberation of a small room or the long sustained reverberation of a large hall.

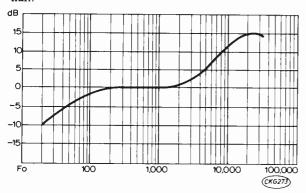


Fig. 3: Frequency response of the spring line driver amplifier.

The Circuit

The complete circuit is given in Fig. 2. The microphone preamplifier Tr1 and Tr2 is suitable for microphones of $200\text{-}600\Omega$, or low impedance microphones with a matching transformer, or with a

crystal microphone. The output from this is taken to a passive mixing network via the microphone gain control VR1. The line level signal input, suitable for any linear signals from about 200mV upward, is taken via the line input gain control VR2 and thence to the passive mixing network R7 and R8.

The mixing network losses are recovered by the amplifier Tr3 and the output from this is taken via a passive network to a) the output emitter follower stage Tr4 and b) to the reverb unit driving amplifier Tr5, Tr6, Tr7 and Tr8. This amplifier delivers around 1W at full reverb i.e., with the reverb control VR3 at maximum. The amplifier also has a special frequency response as shown in Fig. 3 to suit the response require-

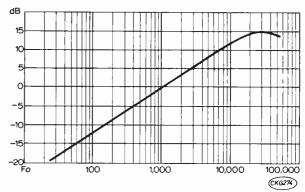


Fig. 4: Frequency response curve of the reverb output amplifier.

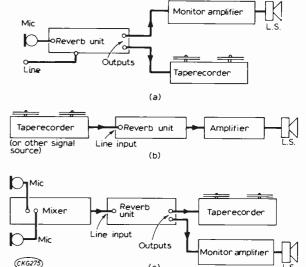
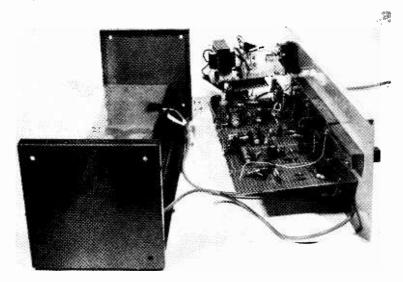


Fig. 5: Ways of using the spring line reverb unit.

(c)



A view of the prototype showing the delay line at the bottom of the case with the electronic components mounted on the back of the front panel.

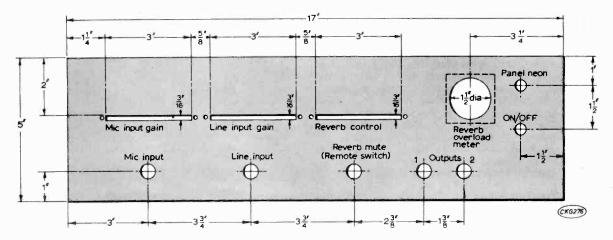


Fig. 6: Details of the front panel component positions and the drilling dimensions

ments of the spring line driving transducer.

The output from the driving amplifier is taken to the spring line unit socket marked 'Input' and also to the reverb overload meter circuit R37, D1, C20 and the 50μ A meter. The use of the meter will be explained later.

The output from the spring line unit, from the socket marked 'Output', is taken to the amplifier Tr9 which also has a special frequency response as shown in Fig. 4. The signal from this amplifier is then taken to the emitter follower output stage Tr4.

The output from Tr4 is split to two output sockets, both of which can be used to feed a tape recorder and/or amplifier, or two tape recorders, or amplifiers as required and as shown in Fig. 5.

The reverb unit is mains operated by the trans-

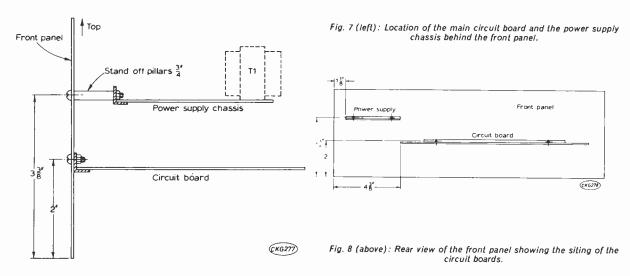
former T1, bridge rectifier BR1 and smoothing network C13, R23, and C12. The supply rail at C13 is 18V positive and this goes direct to the spring line driving amplifier. Further smoothing is needed for the signal amplifier stages, hence the addition of R23 and C12 from which the supply rail is 15V positive. Note that the transformer MT280 has two 6V secondary windings which are connected in series to provide 12V to the bridge rectifier BR1.

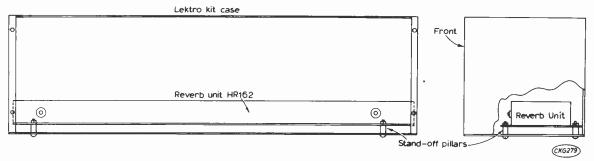
Construction

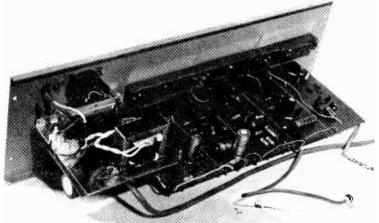
The spring line reverb unit is a type HR162 (Henry's Radio Limited) and is 16^{7}_{8} in long. This will just fit into a Lektrokit case (17in long) in which the prototype unit as shown in the photos was built

★ components list

Resist	ors						onductors				
R1	680Ω	R16	1kΩ	R31	Ω 089		BC109		BC108	Tr9 BC109	
R2	150kΩ	R17	22kΩ	R32	56Ω		BC109		A C128	BR1 BY164 or	
R3	150kΩ	R18	10kΩ	R33	1kΩ		BC109		A C176	similar silicor	
R4	390kΩ	R19	4 · 7kΩ	R34	0.5Ω	Tr4	BC109	Tr8	AC128	bridge rectifie	
R5	1kΩ	R20	1kΩ	R35	0.5Ω					(40V, 1·5A)	
R6	10kΩ	R21	4 · 7kΩ	R36	56Ω						
R7	15kΩ	R22	4 · 7kΩ	R37	22kΩ	Misco	llaneous				
R8	15kΩ	R23	68002	R38	12kΩ			VR3 Fa	ch 10kO 1	log. Slider types	
R9	12kΩ	R24	47kΩ	R39	120kΩ	¥11.1,	VIVE and		gle RR11	og. Olider type.	
R10	120kΩ	R25	15kΩ	R40	1kΩ	PR1			0Ω preset c	arbon tyne	
R11	10kΩ	R26	120kΩ	R41	220Ω		ng Line Un			(Henry's Radio	
R12	1kΩ	R27	39kΩ	R42	10kΩ	Spin	ig Line On	Lt		(Holli) o Hadis	
R13	100Ω	R28	22Ω	R43	1kΩ	Tran	sformer T1			6V+6V second	
R14	56kΩ	R29	47kΩ	R44	$68k\Omega$		3.0111101 1			agle Type MT280	
R15	10kΩ	R30	390Ω			Neor	1			rototype is Eagle	
All resistors ‡W, 5 per cent tolerance except R34				14001			R30)	nototype is Eag.			
and R35 which are 1W.				On-off switch SW1 Toggle type							
					Overload Meter M1 50 µA, Eagle type MR2P						
Capac		040	400 5	040	400 5		Sockets		off required		
C1	5μF		100µF		100μF		no Plugs		off required		
C2	100pF	C11	100μF	C20	10μF	1 1101	io i lugo		on required		
C3	10μF	C12	250μF	C21	250µF						
C4	5/4F	C13	2000μF	C22	5μF	Case	9			5 x 5in. Made up	
C5	5μF	C14		C23	1μF					panel (1) LK401	
C6	5μF	C15	6800pF	C24	0·1μF					(2) LK311, Top	
C7	50μF	C16	25µF	C25	5μ F					ottom Plates (3)	
C8	100μF	C17	250µF	C26	6800pF				(501.		
C9	5μF		4700pF							n Home Radio	
All values above 1μ F are electrolytic types and should have a working voltage of at least 25V,					Circ	uit Board		Plain Veroboard, 0·15in. matrix, see text			
except C13 which should be at least 35V. Values below 1µF can be any type, Mylar, ceramic etc.					Alue	ninium	16	16s.w.g. (for heatsink and power supply chassis); } x } in, angle,			







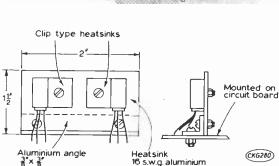


Fig. 10: Details of the heatsink for Tr7 and Tr8.

Fig. 9 (above): Siting of the spring line unit within the case.

On the left is shown a view of the prototype illustrating the siting of the component boards.

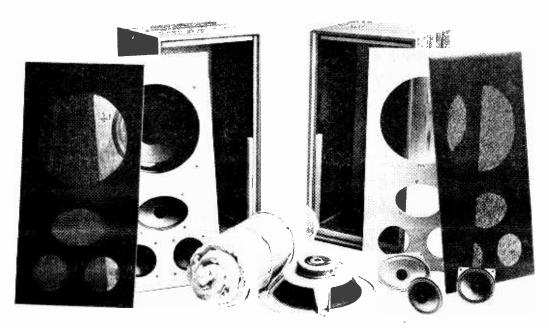
(see components list for details of the Lektrokit case which is 17×5 ×5in). A similar metal case of the same dimensions could of course be used.

Details for drilling the front panel to accommodate the slider type controls and signal sockets etc. are given in Fig. 6. The size of the hole for the overload meter will depend on the meter used but it must be a $50\mu A$ type.

The circuit board containing all the amplifier stages is mounted behind the front panel as shown in Figs. 7 and 8. The power supply is assembled as a separate chassis and this is mounted as shown. The spring line unit is mounted at the bottom and rear of the case and raised slightly on stand-off pillars as shown in Fig. 9 and with the output and input sockets facing towards the front.

The next stage is the circuit board which is $10^{7}_8 \times 5$ in. and is mounted on the front panel by means of $^{3}_8 \times ^{3}_8$ in. aluminium angle and the stand-off pillars as in Fig. 7. Note that Tr7 and Tr8 must be mounted on a heatsink as in Fig. 10 which is made of 16 s.w.g. aluminium to the dimensions given.

TO BE CONTINUED



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Crossover No	etworks		CT10/8	Horn	2 40
CN23	2 way	110	CT10/16	Horn	2 40
CN28	2 way	110	HT15	Horn	3 50
CN216	2 way	1 10	MHT10	Horn	3 60
SN75	2 way	2 20	HT21	Horn	4 20
FF5	3 way	2 60	DT33	Dome	5 20

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When completed, the receiver has three wave-bands—long waves, medium waves and bandspreading of the h.f. end of the m.w. band (also sometimes termed Luxembourg Bandspread). However, though the switching associated with the provision of such bands is not very complicated, it does seem that it is sometimes a source of difficulty or wiring errors

during building. So the receiver is first constructed with one band coverage, tuning m.w. only. When it has been found to give proper results in this form, the few additional items required to add bandspread tuning and l.w. coverage are incorporated. It would, in fact, be easy to limit coverage to l.w. and m.w. only, if the bandspread range is not required, or even to omit l.w. where this band is not wanted.

CIRCUIT

Fig. 1 is the circuit as constructed for one waveband, the m.w. ferrite rod winding L1 tuned by VC1, and the oscillator coil L3 tuned by VC2. VC1/2 is

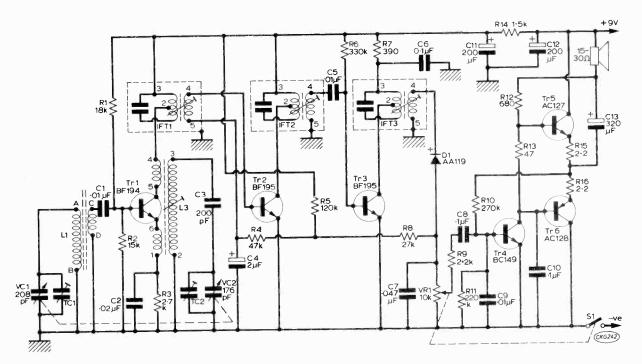
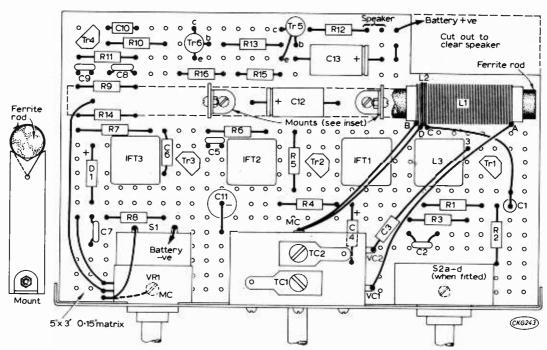
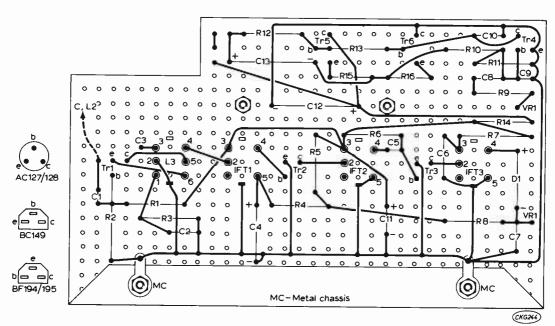


Fig: 1. Circuit of the basic receiver for normal medium wave reception. The additional bandswitching circuit is shown in Fig. 4.



▲Fig. 2: Plan view of components mounted on top of circuit board with details of the mounting for the ferrite rod aerial. NOTE: Speaker should be shown connected to centre and RIGHT-HAND pin.

▼Fig. 3: Wiring on reverse side of circuit board together with transistor lead-out connections.



a ganged capacitor with the trimmers TC1 and TC2 incorporated.

L2 is the m.w. base coupling winding for the oscillator/mixer Tr1. Tr2 and Tr3 are the IF amplifiers using BF195's. Diode D1 is the detector or demodulator which also provides a.g.c. bias for Tr2. The collector circuit of Tr3 is decoupled by C6 and R7. Because of the very high gain of these transistors, it may be necessary to take certain steps to avoid squegging or instability, as described later. This depends to some extent on individual transistor characteristics.

VR1 is the volume control feeding Tr4. Tr5 and Tr6 are used in a directly coupled driver-output arrangement which will be found to give good results and plenty of volume with a 15Ω or 18Ω speaker.

CONSTRUCTION

The controls are mounted on a $5\times2in$. "universal chassis" flanged member, Fig. 2. The top flange is cut off, to clear TC1 and the variable capacitor. When mounting this item, with its spindle through a clearance hole, it is necessary to use very short bolts, or

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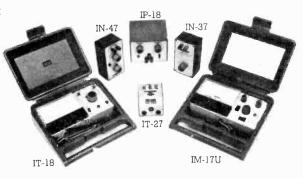
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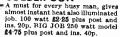
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6for 21in. 13 w. ministure tubes £1. Postage on
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Designed to operate transistor sets and amplifiers. Adjustable output 6v., 9v., 12 volts for up to 500mA (class B working). Takes the place of any of the following batteries: PP1, PP3, PP4, PP6, PP7, PP9 and others. Kit comprises: mains transformer rectifier, smoothing and load resistor, condensers and instructions. Real snip at only 83p, plus 20p poetage.

MULLARD AUDIO AMPLIFIERS All in module form, each ready built complete with heat sinks and connection tags, data sup-

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which are infinitely adjustable for position.
They are also arrange: to allow 2 operations per switch per rotation. There are 16
changeover micro switches each of 10 amp
type operated by the trips thus 15 circuits may
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machines, Display lighting animated and signs, Signalling, etc. Price from
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Where postage is not stated then orders over £5 are post free. Below £5 add 20p. Semiconductors add 5p post. Over £1 post free. S.A.E, with enquiries please.

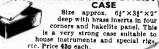
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for \$22. Ditto as above but for printed
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balance kit. Kleven parts, including candle, one concave lens, one convex lens, stage and slit frame, etc. Watch light rays bend as they pass through different lenses. Ka3 Water Pump Kit. Thirteen parts. Top of pump is transparent so that operating parts may be observed. Small parts are brightly colourer to be seen caslly while working. Three types of pump may be made: Lift pump. Force Pump and Force Pump an

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Explosion religious of which kas Bell Kit, Eight parts, including bell and push button switch, Build a complete electric bell and see how the hammer is triggered to make the see horbell rin

KA10 Morse Key buzzer and bell kit. 25 part kit sy to construct, simple to operate.

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* components list

Resistors				
R1 18kΩ	R7	390Ω	R13	47Ω
R2 15kΩ	R8	27kΩ		1 · 5kΩ
R3 2·7kΩ	R9	2·2kΩ		$2 \cdot 2\Omega$
R4 47kΩ		270kΩ		2·2Ω
R5 120kΩ		220kΩ	1110	
R6 330kΩ		680Ω	ΔII 4	W 5%
VR1 10k Ω log.				

Capacitors

C1	0.01#£	Ç9	υ υ ιμε
C2	0 · 02µF	C10	0 · 1μF
C3	200pF 5%	C11	200μF 12V
	2μF 6V	C12	200μF 12V
C5	0.01µF	C13	320μF 6V
	0·1μF	C14	150pF 5%
	0·047μF	C15	30pF SM
	0·1μF		•

VC1-2 208+176pF gang with slow motion and trimmers (TC1-2) (Jackson OO)
TC3-4-5 60pF compression trimmers.

0.04..E

Semiconductors

Tr1	BF194	Tr4	BC149	D1	AA119
Tr2	BF195	Tr5	A C127		
Tr3	BF195	Tr6	AC128		

Inductors

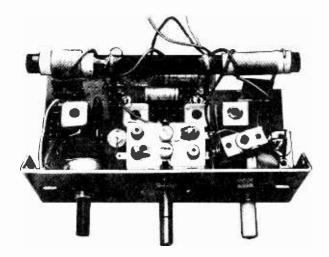
Ferrite rod aerial with m.w. and l.w. windings (Denco MW/LW/5FR) L3 Oscillator (TOC1 Denco) IFT1-2 (IFT13 Denco) IFT3 (IFT14 Denco)

Miscellaneous

Speaker 2½in. dia. 15 to 30Ω. S2, 4 pole 3 way wafer switch. Veroboard 5 x 3in. 0·15in. matrix. Universal chassis members 6 x 3in. (2 off), 5 x 2in. (1) (Home Radio). Knobs, materials for case.

extra washers, to avoid shorting or damaging the front section. Two bolts through the circuit board and remaining flange hold these together, soldering tags being fitted for chassis connections, as in Figs. 2 and 3.

If the 0.15in. matrix board has 19 rows of holes one way and 32 rows the other, this will allow parts to be placed exactly as shown. The board is 5×3 in.



View of completed chassis incorporating band switching.

Prepare it in advance by drilling holes for the bolts and for the pins and screening can tags of L3 and the i.f.t's. A very small round file can be very useful here if any holes are not quite correctly placed. Note that L3 has closer spacing between pins 1-2 and 5-6, and must be fitted as in Fig. 3.

A bare wire, such as 22 s.w.g. tinned copper, is soldered between the tags, as in Fig. 3, for earth returns. One screening can tag of L3 and each i.f.t. is connected to this.

Resistors and capacitors are inserted into the board as in Fig. 2, with their wire ends projecting. The wires are then bent over and connected as in Fig. 3, excess being snipped off. Insulated sleeving should be put on all leads which may touch each other or any metal part. In some places connecting leads have to be soldered on to reach to other components.

It is quite helpful to use black sleeving for the negative leads, red for positive wires and some other colour for other connections, as this makes it easier to check, if necessary.

Leads for the speaker and battery positive can be soldered to pins in these holes, or the wires may project a little and be used as soldering points.

TRANSISTORS

If Tr1, Tr2, Tr3 and Tr4 are placed as in Fig. 2, the pins will come as in Fig. 3. Leads to these pins should be soldered carefully so as to obtain proper joints, yet avoiding lengthy heating. Note that the BC149 pin connections differ from the BF194/195 types.

Place Tr5 and Tr6 so that their lead-out wires emerge as in Fig. 3. These transistors can be left with ¹₂in. or so of lead above the circuit board. Fit diode D1 with the polarity correct as shown.

FERRITE AERIAL

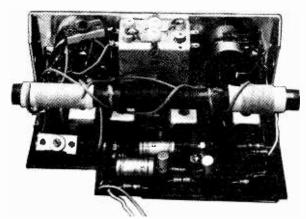
Two strips of paxolin or other insulating material are cut as in Fig. 2 to raise the rod about 1^{1} ₂in. from the circuit board. These strips are fixed with brackets, and the rod is held with string, thread or elastic through holes and round the rod.

For m.w., the m.w. section only is required, and is placed with the coupling winding L2 away from the rod end, Fig. 2. A and B are the ends of the long winding L1, and C and D are the ends of the smaller winding L2. Thin sleeving is put on the wires and enough slack left to allow the windings to be slid along the rod.

IF ALIGNMENT

The speaker should be connected to the points as shown. Core settings of the i.f.t's as supplied should be found to be approximately correct, but small adjustments are made for the best results, using a Denco TT5 tool or similar means of rotating the cores.

If a signal generator is available, tune it to 470kHz and couple it to the base of Trl through a small isolating capacitor, or place an insulated lead near Cl. The i.f.t. cores are then adjusted for best volume. If this is done by ear, set VR1 at about maximum volume, and keep the generator signal input down so that the actual volume is quite low.



Rear view of receiver. Cut-out in veroboard to clear particular speaker used.

When no signal generator is available, tune in properly any stable transmission and adjust the cores for best results. Repeat with a weak signal.

Once the i.f.t. cores are adjusted for best results, they do not need any further attention and should be left.

MW ALIGNMENT

If a signal generator is used, band coverage can be set by adjusting TC2 at the h.f. end of the band and the core of L3 at the l.f. end of the band. A small adjustment of L3 core has considerable influence on band coverage.

A signal is then tuned in with VC1/2 nearly open and TC1 adjusted for best reception. VC1/2 is then adjusted to tune in a signal near the l.f. end of the band and the m.w. winding on the rod moved along to give best results. As these adjustments influence each other they are repeated until no further improvement can be obtained. If alignment is correct adjustment of TC1 or the winding on the rod should not give any important increase in sensitivity anywhere throughout the tuning range.

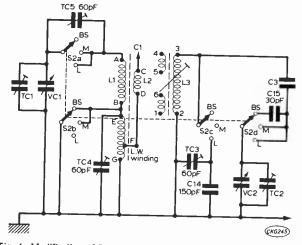


Fig. 4: Modification of Fig. 1 to permit reception on long waves plus bandspreading of h.f. end of medium wave band.

LW AND BANDSPREAD

Long waves and bandspreading are added by changing the aerial and oscillator circuit to that shown in Fig. 4. The 4-pole 3-way switch S2 has three positions BS, MW and LW.

In the BS position, S2d places C15 in series with C3, reducing the coverage obtained with the full swing of VC2. S2a leaves TC5 in circuit, to obtain a suitably reduced tuning coverage in the aerial circuit.

For m.w. reception, C15 and TC5 are out of circuit and S2b shorts the l.w. section of the aerial, giving m.w tuning as for the circuit in Fig. 1.

When l.w. reception is wanted, S2b places the m.w. and l.w. aerial windings in series, while S2c brings in the additional capacitor C14, with parallel trimmer TC3, to shift the oscillator coil coverage.

The tuning ranges obtained were as follows:

BS—1800-1425kHz. MW—1525-560kHz. LW—280-170kHz.

Any departure from this is not important, however, provided the circuits are adjusted for best results, as described.

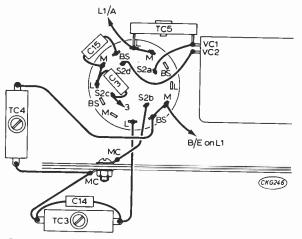


Fig. 5: Wiring of the components around the bandswitch shown in Fig. 4.

Fig. 5 shows the wiring of the bandswitch. TC5 is soldered directly from VC1 to the switch tag, while TC3 and TC4 occupy the space left on the circuit board.

Place the l.w. section of the aerial winding so that the end having two leads is towards the end of the rod. The twisted connection, consisting of two wires, is tapping F and D of the m.w. section is soldered to this. The single wire at the rod end, point G, is soldered to tag MC on VC1/2, Fig. 2. Ends B and E are soldered together and to the tags of S2b.

Note that the tapping F is electrically near point G. There are many more turns from E to F, than from F to G. The direction in which the turns are wound from A, through to B, to E and through to G, is the same. This will be so if the windings are connected as described. If one winding is reversed, I.w. alignment is impossible. Turns C to D and F to G should also be in the same direction throughout. If necessary leads C and D are reversed to secure this.

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H4 2	250	Mixed Resistors. Approx. quantity counted by weight	50
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AF186	0 - 20	2N1304-5	0 - 17
AF139	0.30	2N1306-7	0 - 20
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BC169	0.12	OC20	0 · 40
BF194	0.15	OC23	0 - 25
bF274	0 - 20	OC25	0 - 25
BFY50	0 - 15	OC26	0 · 25 0 · 30
BSY25	0.13	OC28	0 30
BSY26	0 · 13	OC35	0 - 25
LSY27	0 · 13	OC36	0-35
F2A5B	0 - 13	AD149	0 75
8SY29	0 - 13	AUY10 25034	0.75
65Y95A	0 - 10	25034 2N3055	0.40
OC41	0 - 15		0.40
O C44	0 13	Diodes	
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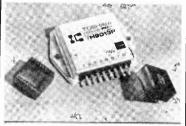
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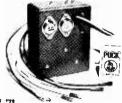
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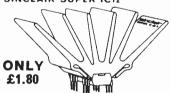
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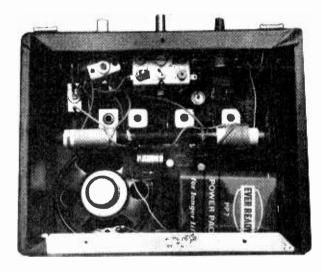
Mail order only. Postage 37p per order on orders containing one or more items with prices marked with an*. Postage on other orders 10p.

3-BAND ALIGNMENT

With the switch at MW align as previously described. A check can be made that coverage is about 560-1525kHz. If very much out, adjust L3 core and TC2 to correct this. Adjust TC1 at about 1450kHz and L1 on the rod, at about 625kHz for best results.

With the switch at BS, C15 is fixed, so no change of coverage can be arranged, except by altering coverage on m.w. Adjust TC5 at about 1500kHz for best reception. This adjustment should be found to remain substantially correct through this range but TC5 may be peaked on Luxembourg if wished.

With the switch at LW, adjust TC3 so that the BBC on 200kHz is about the middle of the band. TC4 may then be peaked at about 250kHz and the long wave aerial winding slid along the rod, at about 180kHz, for best results. The position of the core of L3 greatly influences l.w. coverage, so it is necessary either to set this for a m.w. coverage about as mentioned, or to adjust L3 until l.w. reception is suitable, afterwards moving L1 slightly to compensate so that m.w. reception is also satisfactory. All the adjustments can be repeated, in the order given, until no further improvement is possible.



Rear view of the completed receiver showing placement of the battery and speaker.

INSTABILITY

Commercially made receivers may employ individually selected sets of transistors or make provision for changes to some values, to suit individual transistors, in order to avoid instability.

In this circuit, where oscillation may arise due to particularly "good" transistors, a cure can be effected by adding a 14W resistor between pin 4 of L3, and pin 2 of i.f.t.l. The lowest value which prevents oscillation is used, but it is not critical. A few spare resistors, such as 220Ω , 470Ω , 680Ω or $1k\Omega$ can easily be tried. Should the trouble seem difficult to cure, a resistor of some 47 or 100Ω or so may be placed between point C of L2 and C1. Alternatively, a turn or so may be removed from L2, but this is less easy and difficult to replace.

COMPLETION

The completed receiver, with speaker and battery, can be fitted in any suitable wooden or plastic case. The case illustrated is quite easily made and is approximately 5in. high, 6^3 ₈in. wide, and 3^3 ₈in. deep.

Top and bottom consist of "universal chassis" flanged members 6×3in. The top is drilled so that the bushes of the bandswitch, VR1 and VC1/2 spindle, can pass through. Washers or spare nuts are put on the bushes of the switch and VR1 and the receiver is then secured in place by further nuts on these bushes, when the case is finished.

The case sides are r_0^3 in. 3-ply, each 5×3 in. They are fixed to the top and bottom by 6BA countersunk bolts through the end flanges. Front and back are

similar 3-ply and are $6^{3}_{8} \times 5$ in.

An aperture is cut in the front panel to match the position and diameter of the speaker cone. Adhesive is spread on the case edges and the front fixed by countersunk bolts through the flanges with a few small panel pins driven through into the wooden sides. The back is held with self-tapping screws into the back flanges.

A piece of speaker material about $5^{1}2 \times 3^{1}2$ in. is glued across the front. The front was then covered with self-adhesive material, which can be obtained in many colours and styles. Top and sides are covered in the same way. Strips are cemented on to cover the junction between the speaker gauze and the case covering material.

TELEVISION

NOVEMBER ISSUE

LARGE-SCREEN TV

For large-screen public TV displays it is necessary to adopt some system other than the simple direct viewing of a c.r.t. screen. Many systems have been tried and used and some are suitable for colour TV displays. Development work is still in progress on some techniques. A particularly interesting large-screen colour TV system was used at EXPO 70: this employed mechanical scanning and, as the light source, gas lasers. We shall be looking at this and other systems—such as Eidophor techniques—this month.

RENOVATING THE RENTALS

Caleb Bradley tackles another colour chassis widely used for first-generation rental purposes, the Bush-Murphy CTV25/CV2510 series.

COLOUR RECEIVER CIRCUITS

Gordon King resumes his series, investigating vertical shift techniques and line oscillator circuits, both valve and transistor.

ON SALE OCTOBER 16

SICON SIGNATURE SINGLE SINGLE

LTHOUGH desirable for reliability in operation, silicon power transistors were, until recently, rather costly devices. They can now be obtained at prices comparable with germanium types and for this particular project the Newmarket type NKT 0033 have been chosen for the output stage, together with other small signal silicon types for the driver and input stages and for a recommended signal preamplifier.

The power amplifier, the circuit shown in Fig. 1,

will deliver 8W r.m.s. into an 8Ω load for a THD factor of not greater than 0.2 per cent. The input sensitivity can be varied between 250mV and 1V depending on the value of Rb which sets the amount of negative feedback. The output pair are controlled against temperature rise by diodes D1 and D2 although it is most unlikely that any problem in this direction would occur, providing the heatsink for the output pair is of sufficient area.

The signal input is taken via Tr1 to the initial driver Tr2 which is in series with the bias pre-set PR1 for setting the d.c. condition of the driver and output stages. The drivers (Tr3 and Tr4) are a complementary NPN-PNP pair. Overall feedback is controlled

Fig. 1: The circuit of the main amplifier which uses all silicon transistors and gives an output of 8W r.m.s.

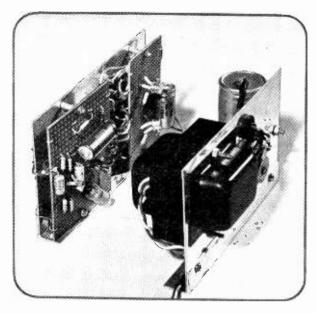
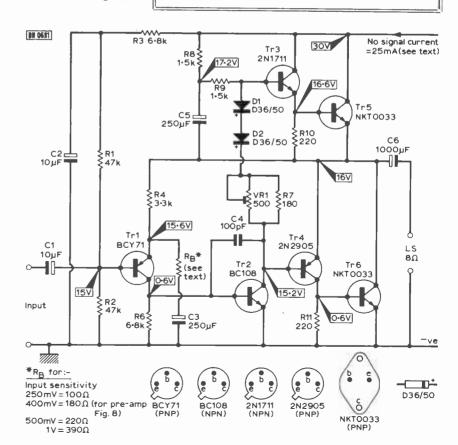


Table 1. Main Amp Specification

Frequency Response Max. Power Output Signal-to-Noise Distortion at 1kHz

30Hz to 30kHz \pm 1dB 8W into 8 Ω load Better than -90dB 8W less than 0·2 per cent 1W less than 0·3 per cent 500mW less than 0·4 per cent



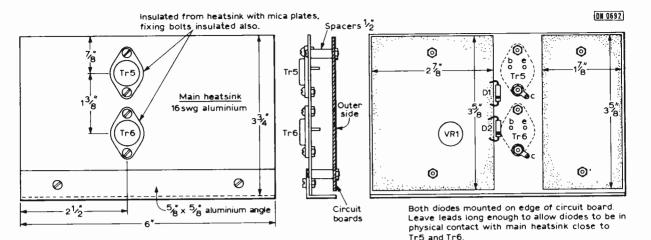


Fig. 2: (a) The heatsink for the output transistors. (b) An edge view showing the mounting of the circuit board. (c) The siting of the two circuit boards.

by the value of Rb and this also determines the required input sensitivity. The table shows the overall performance obtained with the prototype.

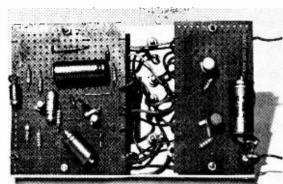
The amplifier is stable on loads down to 4Ω and will operate at reduced power output (approx 6W) into a 15Ω load. The quiescent (no signal) current is approximately 25mA for a supply rail of 30V. Total current at maximum power output (8 Ω load) is about 400mA at which the supply voltage falls to about 26V to 27V.

CONSTRUCTION

The construction and assembly of the prototype

* components list - Main Amplifier

_				
	R3 R4 R5 R6 All r	47kΩ 6·8kΩ 3·3kΩ Rb—see text 6·8kΩ esistors ‡W, 5	R10 R11	1·5k Ω 1·5k Ω 220 Ω 220 Ω
	Capac C1 C2 C3	10µF 20V m	in.	C4 100pF C5 250µF 10V min. C6 1000µF 35V min.
	Tr1 Tr2. Tr3 Tr4 The Elec Gree	trovalue Ltd.,	28 S urrey	Tr5 NKT0033 Tr6 NKT0033 D1 D36/50 or IN4001 D2 D36/30 or 1N4001 are available from either it. Judes Road, Englefield or Arrow Electronics, 7 pod, Essex.
	Trar Brid	r Supply nsformer T1 ige Rectifier M acitor C1	(H R1 T; (H	10V/17W Type TS 18/2 Henry's Radio) ype LT119, 30V, 1Α Henry's Radio) 100μF 50V



A view of the prototype illustrating the arrangement shown in Fig. 2c.

is shown in Fig. 2. The heatsink must not be less than the area specified, even if an alternative layout and construction is adopted. The assembly shown makes a quite compact unit with the two circuit boards backed onto the main heatsink.

The layout and wiring are shown in Fig. 3a but note that the components are as seen from the outer side of the boards and the wiring from the inner side. The power transistors Tr5 and Tr6 must of course be completely insulated from the heatsink with mica spacers and fixing bolt insulators as shown in Fig. 3b. Note also that the two diodes D1 and D2 must be wired so that they can be pushed down to make physical contact with the heatsink. The complementary pair drivers Tr3 and Tr4 do not require heatsinks.

POWER SUPPLY

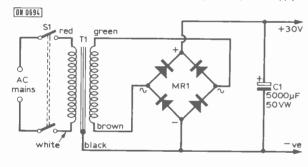
The circuit and layout are given in Figs. 4 and 5 respectively. The construction of the prototype is on the same lines as the power module with the transformer and rectifier etc. mounted on a piece of aluminium $6in \times 3^3 in$. The transformer and rectifier specified will only supply sufficient current for one power amplifier module and a signal amplifier. If two power modules are to be used for stereo operation then the power supply must be able to deliver at least 1A with a maintained voltage (under maximum

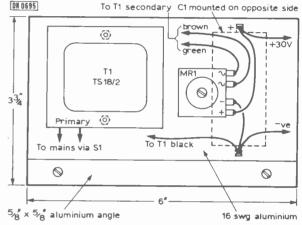
Components as seen from outer side of circuit boards. Wiring as seen from opposite side. DN 0693 Circuit board A Circuit board B Mica insulating Heatsink washer \^^ +30V ≷R8 ≷R10 Input ixing bolt Tr5 and Tr6 insulating Tr6 eyelets C6 C2 ₹ R Оh C3 LS R11 Note: Tr5 and Tr6 completely insulated from Common earth tag Collector connection through solder tag heatsink. Collector connection via solder bolted to and fixing bolt to case of transistor tag under insulated fixing bolt. main heatsink (a) (b)

Fig. 3: (a) The component wiring of the main amplifier.
(b) The mounting of the output transistors.

Fig. 4 (below): The circuit of the power supply.

Fig. 5 (right): The wiring of the power supply.





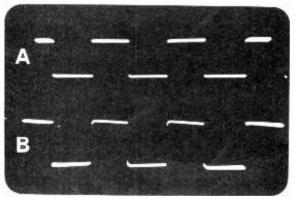


Fig. 6: Square wave results on the main amplifier.

B ////

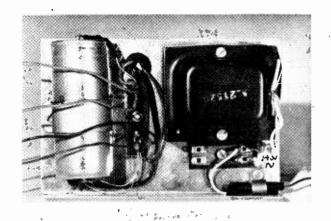
Fig. 7: Oscillogram showing symmetrical clipping.

power load) of 26-27V. The same rectifier may be used, as this is rated for 1A at 30V but a transformer capable of supplying the higher current will be required.

POWER AMPLIFIER PERFORMANCE

Before connecting the supply voltage, set PRI to midway travel. Connect a milliamp meter in series with the positive line and reset VRI to obtain a reading of 25mA. At maximum power, (just before

clipping) the total current should not exceed 400mA. The input sensitivity can be adjusted by the value of Rb and typical settings for this are given in Fig. 1. The overall performance should be as given earlier but square-wave tests will further prove the performance. The result of the usual 1kHz square-wave test on the prototype is shown in Fig. 6 in which the input is the oscillogram A and the output from the amplifier is oscillogram B. Symmetrical clipping should be obtained at just over maximum (sine-



The prototype power supply.

Table 2. Preamplifier Specification

Frequency Response

Tone Controls

Distortion Factor

Signal-to-Noise Inputs (for 400mV out) Linear: ± 1dB, 10Hz to 100kHz

Tape Head: CCIR plus treble lift

Mag.P.U.: RIAA

Bass ± 14dB at 50Hz Treble ± 12dB at 10kHz Less than 0.2 per cent

-60dB ref. 400mV output Tape Head: 68kΩ, 5-10mV Mag. P.U.: $56k\Omega$, 5mV

Cer. P.U.: 820kΩ, 80mV Radio: 200kΩ, 80mV

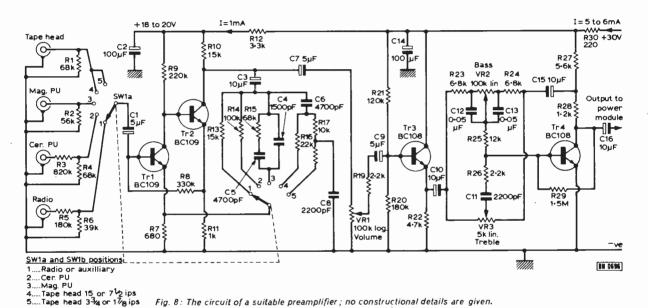


Fig. 8: The circuit of a suitable preamplifier; no constructional details are given.

wave) power output as shown in Fig. 7 in which A is a 1kHz sine-wave input and B the clipped output.

The average d.c. potentials that should be obtained with a 20,000 ohms per volt meter at various points in the amplifier are marked on the circuit; all with respect to common negative.

PREAMPLIFIER

A circuit for a suitable preamplifier is given in Fig. 8 and this caters for magnetic or ceramic pick-up cartridges, tape head and audio. The radio input can be used for tape replay from a unit with built-in preamp. The performance of the preamplifier is shown in Table 2.

The overall gain has been adjusted to provide 400mV output to match the input requirements of the power amplifier with Rb (as in Fig. 1) a value of 180Ω . The input circuit Tr1/Tr2 is quite conventional with selective feedback to provide equalisation for tape head and/or magnetic pick-up. The tape head response has been adjusted to comply with the recording characteristic most frequently used on domestic tape recorders; one which has less treble

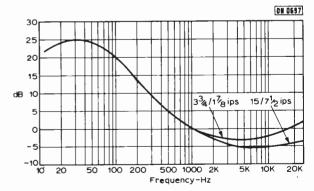


Fig. 9: Preamplifier response when switched to Tape Head.

lift than the original C.C.I.R. characteristic. The replay response obtained with the preamplifier is shown in Fig. 9 and provides what is effectively treble lift towards 20kHz but which is adjusted accordingly for tape speeds of 15 and 7¹₂ or 3³₄ and

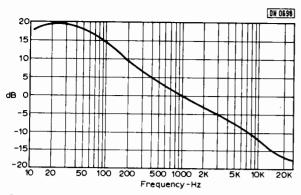


Fig. 10: The preamp response on Mag. P.U. showing R.I.A.A. curve.

* components list - Preamplifier

RESISTORS		
R1 68kΩ		22 kΩ
R2 $56k\Omega$	R17	10kΩ
R3 820k Ω		12k Ω
R4 $68k\Omega$		2·2kΩ
R5 180 k Ω	R20	
R6 39k Ω		120kΩ
R7 680Ω	R22	4·7kΩ
R8 330k Ω	R23	6·8kΩ
R9 220kΩ	R24	6·8kΩ
R10 15k Ω	R25	12kΩ
R11 $1k\Omega$	R26	2·2kΩ
R12 3·3kΩ	R27	
R13 15k Ω	R28	1·2kΩ
R14 100k Ω		1·5MΩ
R15 $68k\Omega$		
All resistors ‡W, 5 p	er cent	tvpes
VR1 100k Ω log. pot.	(volume	
VR2 100kΩ lin. pot.	(bass)	•
VR3 5k Ω lin. pot. (tr	eble)	
Capacitors		
C1 5µF 25V min.	C9	5μF 25V min.
C2 100µF 25V min.	Č1	
C3 10µF 25V min.	Č1	
C4 1500pF	Č1	
C5 4700pF	C1	
C6 4700pF	Č1	
C7 5µF 25V min.	C1:	
C8 2200pF	C1	
Semiconductors		,
Tr1 BC109	Tra	BC108
Tr2 BC109	Tr	
SW1 Double pole, f		
polo, [··· way	iotally officell

178 i.p.s. The response for magnetic pick-up is to R.I.A.A. characteristic as shown in Fig. 10.

The output from the first stage (Tr1/Tr2) is taken via the volume control to an emitter follower Tr3 and then via the active (feedback) tone control network to Tr4.

Care should be taken to prevent hum pick-up in the first stages, particularly in view of the high gain when the tape head or magnetic p.u. inputs are in use. The preamplifier can of course be duplicated for stereo use, in which case a balance control could be included between the ouput and the power amplifier. The insertion of a balance control will of course reduce the overall output signal in which case the feedback resistor Rb in the power amplifier should be about 100Ω .

CHARLES MOLLOY

edium ave Column

MAYHEW who lives in Littlehampton asks for information about English broadcasts beamed from abroad. As the medium waves are mainly used for local broadcasting there are not many special broadcasts in English to be heard, but the few known to the writer include Radio Sweden on 1178kHz, Radio Portugal on 755kHz and 1161kHz, Deutschlandfunk (West Germany) on 1268kHz, Radio Pyrgos (Greece) on 1349kHz, Radio Montecarlo 1466kHz.

Radio Sweden is on the air nightly at 2245hr GMT in English and it features a special programme "Sweden Calling DXers" every Tuesday. Radio Portugal transmits the English version of the "Voice of the West" at 2245hrs, its programme for DXers appearing on a Monday. Deutschlandfunk can be heard every evening at 1745hrs and there is a fortnightly DX broadcast on a Wednesday. In spite of rumours of a closedown, Radio Pyrgos is still on 1349kHz. It can be heard in English at 0100hrs on Mondays and its DX broadcast is on the last Friday of the month at midnight. The Vatican Radio's 250kW outlet on 1529kHz is received well in this country and there is a short broadcast in English every evening at 2100hrs. Trans World Radio uses the Radio Montecarlo transmitter on 1466kHz for religious broadcasts which can be heard at 2100hrs.

Harold Emblem of Mirfield in Yorkshire reports reception of the new Voice of America transmitter at Kavalla in Greece which is on 791kHz. The time of reception was 0300hrs GMT. At the moment this station is only using 150kW but it will soon be on full power of 2500kW when it will become the most powerful medium wave broadcaster in Europe. Harold also mentions the Voice of America at Rhodes on 1259kHz; Omdurman, Sudan on 960kHz with the call "Huna Omdurman"; the BBC Central Mediterranean Relay at Point Delimara, Malta on 1511kHz at 0345hrs with Big Ben and "Huna London" at the start of the Arabic service.

An interesting letter comes from Norman Burton who lives in Revesby N.S.W. in Australia. He is a keen MW DXer and he says "This place is good for MW DX. I've copied KFSO (560kHz, 1kW) in San Francisco at midnight, 8500 miles away. At this hour it is 6 a.m. in Frisco as it is across the international dateline." Norman is constructing the PW MW loop antenna which was described in the November 1966 edition of Practical Wireless. Although this issue is now out of print, copies are available in public reference libraries.

Please send logs and information about the medium waves to the author at 132 Segars Lane, Southport PR8 3JG.

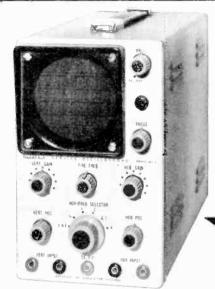
PRACTICAL WIRELESS

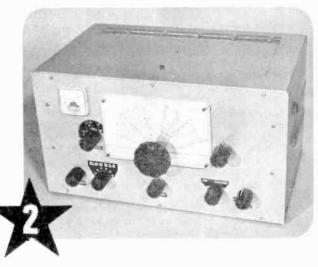
will be at THE AUDIO FAIR, OLYMPIA

A OF THE GREAT FEATURES IN NEXT MONTH'S PW!



10 'scopes to be won!









A great, free competition, open to all readers of P.W. Ten (yes, TEN!) Heathkit Oscilloscope Kits Type OS-2 MUST be won. All you have to do is list eight of the excellent features in their order of importance and you could be the winner of a fabulous 'scope. Watch out for the entry forms which will appear in both the December and January issues.



The Music Maker will be one of the most popular features of the year; a portable, keyless, electronic organ which can be built for a song! The prototype has given the P.W. staff endless hours of fun... and it couldn't be easier: we have arranged for the supply of a special printed circuit board, the case and all unusual components at favourable prices.



AMATEUR-BANDS RADIO

By employing a double conversion circuit, this 6-band receiver provides good adjacent channel selectivity, utilising a crystal filter at 455kHz, while the tuneable first i.f. stage, covering 3.5 to 4MHz, virtually eliminates image interference. The project can be built in easy stages, initially as a receiver for the 80m band, later adding a product detector, BFO and crystal controlled front end converter.

LOUDHAILER & SIREN

You'll certainly be able to stir up the atmosphere with our combined loudhailer and siren... it has all the facilities of commercial models with a lot of extras. Operated from dry batteries inside the main case and using a ready built power module, this is not a complex project. No difficult metalwork is an added attraction to this unusual feature.

PLUS MANY OTHER FEATURES ALL IN THE DECEMBER ISSUE ON SALE 3rd. NOV.

TAKE 2®

JULIAN ANDERSON

HOSE readers of this column who are blessed with offspring will not need reminding that children love noise. Most of the time they will be discouraged but at parties there is no harm in allowing them to let off a little steam. Our project this month makes an excellent, if noisy, children's party game; it is a variation on the time-tested "Hunt the Thimble" game. Instead of a thimble, we use a small "bleeper" which gives off a high-pitched whistle but at very low level so that it can only be heard in quiet surroundings. The idea is to divide the children into two teams; the organiser chooses the hiding place and when this is done the two teams are set loose. They can either concentrate on finding the bleeper or on making so much noise that the other team will not be able to hear it. Teams can of course divide their forces and have "searchers" and "noisemakers"; it doesn't take much imagination to predict which most children will opt for!

The bleeper, if built up as shown, is very robust and should stand up to the rough handling to which it is bound to be subjected. As with all our projects the cost is low, in this case even new components will set you back only about 50p plus the battery cost,

well worth it for a popular party game.

The circuit is shown in Fig. 1. It is a simple multivibrator working at about 2kHz although the exact frequency will depend on the particular transistors used.

The sound output is taken to a very cheap crystal earpiece which is wired in parallel with R4. The circuit works perfectly well with a 1.5V single cell battery but space is a bit limited and a mercury cell Type RM401H may be better. The current drain of the circuit using the specified components is only about $600\mu\text{A}$ —less than a milliamp—and mercury cells will easily give this.

The operation of the multivibrator is not as simple as many people believe but we have described the operation before and no doubt will do so in the future; for this reason we shall not go into the actual theory. All that you need to know is that it is very easy to build and very reliable.

The prototype was built into a 35mm film can. A hole can be made in the lid and the crystal earpiece fitted through it. All the crystal earpieces that I have seen are of the same type, though there may be others. In the commonest (if not only available) type the plastic part which fits into the ear screws off and if the hole is made just large enough to take this thread, the lid of the can may be trapped when the earpiece is reassembled. This is illustrated in Fig. 2.

There is no room for a circuit board but there are very few components and these can be wired to each other as shown in Fig. 3. When this has been completed and tested the components can be wrapped in insulating tape to prevent shorting to the can. A switch is shown in the circuit but once again room precludes the use of a conventional type and a wander plug and socket can be used or even wires twisted together and untwisted after use.

No. 42 "HUNT THE BLEEPER"

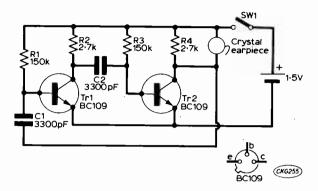


Fig. 1: The circuit is a simple a.f. multivibrator feeding a crystal earpiece.

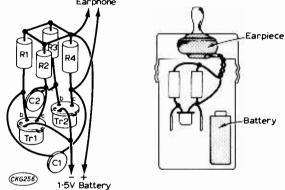
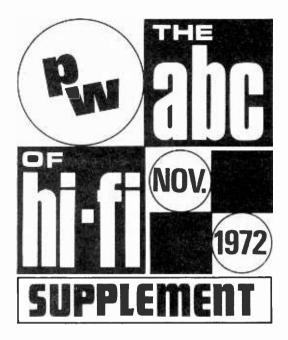


Fig. 2 (left): The components should be wired to each other as shown. Fig. 3 (right): Shows how the components fit into the can.

★ components list

R1	150kΩ 5%, ‡ W	1p
R2	2·7kΩ 5%, ↓W	1p
R3	150kΩ 5%, ↓W	1p
R4	2·7kΩ 5%, ‡W	1p
C1	3300pF ceramic or polystyrene	3p
C2	3300pF ceramic or polystyrene	3p
Tr1	BC109	10p
Tr2	BC109	10p
Crys	tal earpiece	20p
	n film can .	
	•	50p
chan costs	es are those recently advertised and ged. No allowance is made for mining or for postage and packing and the ild be carefully checked before ordering	num order ese points

Contacts to the mercury cell will have to be made by soldering wires directly to it and this operation should be performed with care. I don't know too much about the chemistry of mercury cells but if the soldering operation is not done quickly the cell appears to discharge almost instantly making the case extremely hot and destroying it, so do take care not to heat the cell up too much.



With Hi-Fi now a booming industry the potential purchaser of audio equipment for home entertainment is faced with a mountain of shiny brochures often describing the merits and specifications of equipment in a jargon he finds difficult to understand. This supplement, compiled by Gordon King, the well-known writer on all types of electronic equipment, is aimed at clarifying the more commonly used terms in the field of Hi-Fi.

N EW terminology is being introduced by the progressing art of hi-fi and it is the purpose of this supplement to explain the more important terms used by manufacturers in their specifications as well as some of the established ones found in abundance in the hi-fi literature.

The terms and their definitions, along with sometimes revealing comment, are classified under three headings to which they most appropriately relate. These are Amplifiers, Tuners and Ancillary, the third dealing chiefly with those terms applicable to equipment other than amplifiers and tuners. Sometimes, of course, a term, such as distortion, will be equally applicable under the three headings.

AMPLIFIERS

AMPLITUDE DISTORTION

This distortion is present when the amplitude of the output signal fails to follow exactly any increase or decrease in the amplitude of the input signal. It results from non-linearity of the transfer function and gives rise to harmonic and intermodulation distortion. No amplifier is completely free from the effect because its transfer function is slightly curved.

The nature of the curvature determines the order of the distortion produced, but negative feedback and other artifices used in hi-fi amplifiers minimise the curvature within the dynamic range and hence keep the distortion at a very low level.

ATTACK

This term describes the 'responsiveness' of the amplifier to signals with a fast rise-time, such as produced by percussive sounds of a transient nature. See under Rise-Time.

CHANNEL SEPARATION

This refers to the 'leakage' of signal from one channel of a stereo amplifier into the other channel. The separation is greatest when the signal leakage is the smallest. The leakage signal is commonly expressed as a decibel ratio referred to the full output of the amplifier when the channel in which the leakage signal is measured is correctly terminated both at the input and output. Thus a separation of 40dB means that the voltage of the leakage signal at the output of the channel tested is 100 times below that of the voltage of the signal at the output of the talking channel.

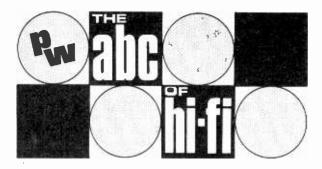
Owing to capacitive effects, the channel separation sometimes falls with increasing frequency, while impairment at the l.f. end can indicate a common impedance coupling between the two channels. One cause of poor separation is the common impedance presented by the wiper of the balance control on the track of the potentiometer.

CLASSA

This implies biasing the valves or transistors to the middle parts of their transfer characteristics so that the device is driven upwards on one half cycle and downwards on the other half cycle. In a push-pull power amplifier, one of the device pair is driven upwards on negative half cycles while its partner is driven downwards, the mode reversing on the positive half cycles. The stage thus draws a constant current at all drive levels within the dynamic range of the amplifier. A class A power amplifier has an efficiency of almost 50 per cent.

CLASS AB

One type of power amplifier of this class is engineered so that at low drive level the stage operates class A, while at increasing drive level the mode changes to class B.



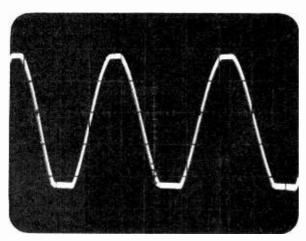
The Mullard Pi-Mode class AB power amplifier design is arranged so that the actual biasing changes from class A to class B with increasing drive.

CLASS B

The majority of transistor amplifiers adopt this class of power amplifier. The biasing is adjusted such that the push-pull transistors (could be valves) operate at a low no-drive current (called quiescent current). When drive is applied the current in one of the pair rises while the partner is pushed into cut-off on one half cycle, the mode reversing on the other half cycle. Maximum efficiency of a class B power amplifier is 78·5 per cent. Current increases with drive.

CLIPPING

This is the term used to express the clipping of the peaks of a waveform when an amplifier is driven beyond its power capacity.



Clipping. The flattening of the tips of a sine wave due to clipping.

COMPLEMENTARY PUSH-PULL

A power amplifier where the output transistors are of complementary polarities (i.e., p-n-p and n-p-n). In some amplifiers of this kind the driver transistors also constitute a complementary pair.

CROSSOVER DISTORTION

The type of distortion resulting from class B push-pull power amplifiers owing to the lack of coincidence of the two transfer characteristics at the crossover point. The effect is reduced by applying a critical value of biasing to optimise the quiescent current (see under Class B and Quiescent Current) and hence 'linearise' the middle portion of the transfer characteristic. The situation is further

improved by heavy negative feedback and by circuit design, such that the crossover distortion from hi-fl amplifiers is very small.

DAMPING FACTOR

This is the ratio of load or speaker impedance to the amplifier's output impedance. Thus the smaller the output impedance the greater the damping factor. The damping factor increases with increase of voltage negative feedback, and with the large amounts of feedback applied to transistor hi-fi amplifiers the source or output impedance can be as low as 0·1 ohm, giving a damping factor of 80 referred to an 8-ohm load.

Such a small value of output impedance Is 'seen' by the speaker as a virtual short-circuit, excluding the resistance of the speaker connecting cable, and this has the effect of damping the cone movement electromagnetically, thereby minimising the speaker output at the resonance frequencies. The bass resonance is important from this respect, so the damping factor should remain desirably high down to 40Hz or less.

DERIVED CENTRE CHANNEL

Some amplifiers are now appearing with a centre channel output derived from the addition of the left and right signals by a matrix. This output can be fed to a single-channel power amplifier operating a speaker for location between the left and right stereo speakers.

DIRECT COUPLING

This refers to interstage couplings or the speaker coupling being direct with no intervening transformer or capacitor. To the speaker it ensures that the damping factor remains high at l.f. (but increasing power supply impedance at l.f. can influence this), while direct coupling generally minimises l.f. phase shift and encourages enhanced bass performance.

DYNAMIC RANGE

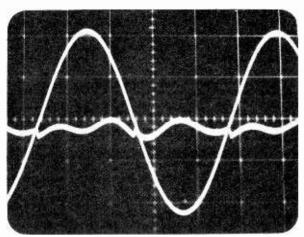
This is the range of reproduction limited on the one hand by the intrinsic noise of the amplifier (see under Noise) and the ambient background noise level of the listening environment and on the other by the power capacity of the amplifier and speaker system. In many cases the dynamic range is limited by the programme signal, but well recorded gramophone records are capable of a dynamic range approaching 60dB, which is 106:1 power ratio (one million).

EQUALISATION

When emphasis is given to a range of frequencies during processing or recording (or transmitting), the correction required to normalise the signals during reproduction is called equalisation. Typical is the RIAA recording characteristic which, apart from ensuring that the recording amplitude is reduced from about 800Hz downwards, boosts the recording level with increasing frequency so that the required de-emphasis (equalisation) during reproduction not only 'flattens' the overall response but also reduces the record noise at the same time.

FEEDBACK

This refers to the feedback of the output signal (or a portion of it) to the input. When the feedback signal is negative it subtracts from the input signal and thus reduces the effective output and when it is positive it adds to the input and increases the effective output. Positive feedback produces sustained oscillation or howl. One example is acoustic feedback where sound from the speaker causes



Crossover Distortion. Artifacts on the distortion residual produced by very mild crossover distortion.

response at a gramophone pickup or microphone, a feedback loop then being completed through the amplifier. Howl or oscillation develops at the frequency corresponding to a major resonance in the loop.

FREQUENCY DISTORTION

This is the unequal amplification of all frequencies over the passband of the amplifier (also see under Frequency Response).

FREQUENCY RESPONSE

The range of frequencies over which an amplifier responds within defined limits of amplification (or signal output). A parameter expressed as 20Hz to 20kHz is the frequency range and is of limited value, while the parameter expressed as 20Hz to 20kHz \pm 1dBisthe frequency response which means that the deviation of amplification from 20Hz to 20kHz is a mere \pm 1dB or \pm 1dB, which is much more meaningful.

HARMONIC DISTORTION

The voltages of harmonics resulting from amplitude distortion expressed as a percentage of the voltage of the fundamental. A common measurement is total harmonic distortion (THD) where the fundamental of a very low distortion sinewave test signal is removed by a steep notch filter, the summed harmonics remaining then being measured as a voltage and expressed as a percentage (or dB value) of the voltage of the fundamental at the required test power of the amplifier.

When measured in this manner the term THD is not really correct since the distortion also has the amplifier noise added to it within the test passband. The correct term is distortion factor.

HEAT SINK

A metal 'radiator' used in transistor amplifiers to conduct the heat away from the power transistors, thereby allowing them to be used for the delivery of greater power.

HIGH-PASS FILTER

A filter which, above a critical frequency, allows the unrestricted passage of high-frequency signals. Reciprocally, a bass-cut filter.

HUM

Spurious signal at the mains frequency (50Hz in the UK) and its harmonics. It is sometimes expressed with noise (see under Noise), so that the total effect is hum and noise.

HUM-LOOP

A condition arising from the connection of two or more 'earths' to an amplifier system whereby circulating currents of low value at mains frequency and harmonics are added to the programme signals, causing hum to appear in the background.

INPUT IMPEDANCE

The effective impedance at the input circuits of an amplifier across which the signal source is connected. In this respect the source is loaded by the input impedance. A load sensitive source is a magnetic pickup, most of which are designed for optimum treble response connected to a load of about $47k\Omega$. A higher resistance will cause a rise in treble, and a lower one a fall in treble.

INTEGRATED AMPLIFIER

An amplifier which embodies in a common housing the preamplifier and control section and the power amplifier. Some early amplifiers of large power were in two separate units, one the control unit with preamplifiers and the other the power amplifier.

INTERMODULATION DISTORTION

Resulting from amplitude distortion (i.e., non-linearity) two or more signals passing through an amplifier intermodulate and hence produce spurious signals, known as intermodulation distortion. Since the distortion components are inharmonious they are subjectively disturbing, more so than low order, even harmonic distortion. Discontinuity in the crossover function of class B amplifiers is responsible for intermodulation distortion and, sometimes, the listening fatigue resulting therefrom. Measurement in based on the voltage of the intermodulation components expressed as a percentage of the voltage of one of the test signals, when two signals are applied to the amplifier in a 4:1 amplitude relationship at a low (100Hz) and high (5kHz) frequency.

LINE INPUT

Input channel of an amplifier designed to accept signal at a given level from a line at a specific impedance, usually 600Ω .

LINE OUTPUT

Output channel of an amplifier designed to deliver signal at a given level to a line at a specific impedance, usually 600 Ω .

LOAD

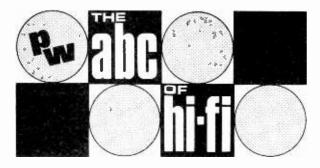
The circuit or transducer (e.g., speaker) connected to the output of an amplifier. The source (e.g., pickup) is loaded by the amplifier's input impedance (see under Input Impedance).

LOUDNESS CONTROL

A control, often the volume control suitably switched, which boosts the bass and to a lesser degree the treble as it is retarded as a means artificially of compensating for the ear's loss of sensitivity at bass and treble frequencies at reducing loudness. A volume control which is permanently arranged to operate in this manner (i.e., loudness effect not switchable) is deprecated by some people and hi-fi enthusiasts.

LOW-PASS FILTER

A filter which, below a critical frequency, allows unrestricted passage of low-frequency signals. Reciprocally, a treble-cut filter.



MATCHING

This commonly refers to the selection of the correct impedance to match the source or load and of the correct sensitivity of the input to match the source signal. If the source yields a signal substantially greater than the input sensitivity of the amplifier the volume control will have to be operated at a low setting and overloading of the input preamplifier may result.

MISMATCH

A condition whereby the coupled impedances inhibit optimum power transference or the source or output signal fails to match the amplifier sensitivity or the load's (i.e., speaker) power capacity.

MUSIC POWER

Sometimes called **dynamic power**, and refers to the power that the amplifier (with both channels working, when stereo) is capable of delivering on music type signal. Output is rated in watts into a specified load value and the test signal is sometimes interrupted sinewave signal (i.e., tone bursts). This method of power expression fails to take account of the power supply impedance and regulation and the efficiency of the power transistor heat sinks. The power obtained is always greater than that measured when the test signal is continuous wave (i.e., steady-state sinewave signal) and with both channels driven together (see under Output Power).

NOISE

Spurious signal resulting from the random electrical charges existing in conductors, components, valves and transistors. It is heard from the loudspeaker as a 'hiss'. Hum (see under Hum) is also sometimes included as general noise, but since the hum is sometimes filtered in certain measurements, hum and noise may be expressed separately.

OUTPUT IMPEDANCE

Effective impedance across the output terminals of an amplifier as reflected across the load (i.e., speaker). This impedance is generally much lower than that of the speaker (see under Damping Factor).

OUTPUT POWER

The maximum amount of power, limited by clipping or a specified value of distortion, the amplifier is capable of delivering to a load of given value (also see under Music Power).

Many transistor amplifiers yield increasing power with reducing load value, and the power is generally given at 4 and 8Ω (sometimes also at 16Ω). The power of stereo amplifiers is given per-channel with one channel only driven and/or with both channels driven together. The almost double demands on the power supply when both channels are driven together reduces the per-channel

power significantly unless the supply is regulated or of a very low impedance when the power amplifiers are class B (see under Class B). However, this does not follow with class A amplifiers (see under Class A) because the power input remains constant within the dynamic range (see under Dynamic Range) of the amplifier.

The most meaningful power rating is that based on continuous wave (i.e., uninterrupted sinewave) signal with both channels driven to capacity together, referred to a specified level of distortion (0.5 per cent or 1 per cent), the power then being calculated by squaring the r.m.s. voltage across the load and dividing by the load value on ohms. This gives the average power per-channel, both running, not the r.m.s. power as often thought and, indeed, specified by many manufacturers!

The peak power on a continuous wave signal is equal to the square of the peak value of the sinewave divided by the load in ohms. Thus when the sinewave passes through zero the power is zero. Hence the average power is equal to half the peak power. The peak power is also equal to the square of the r.m.s. voltage across the load times the square of root-two divided by the load in ohms. Hence half of this, which is then equal to the square of the r.m.s. voltage divided by the load in ohms, is the average power. The power in r.m.s. watts can be obtained by calculation, and it works out to 1·225 times the average power! This is almost 23 per cent higher than the average power. We often wonder whether the spec. writers are aware of this.

OVERLOAD

When an amplifier is driven beyond its capacity the waveform is clipped and overloading and hence heavy distortion results.

OVERLOAD MARGIN

It is desirable to have a margin prior to the onset of overload to avoid clipping on transients. This also enhances the reproduction, giving it a 'smoother' quality in many cases.

POWER BANDWIDTH

The bandwidth over which the full power does not drop below a specified value. The l.f. and h.f. points where the power is half that at 1kHz (i.e., full power) are often given. This is the half-power bandwidth. Expression is commonly such as '20Hz to 35kHz -3dB', the -3dB indicating a power drop to half the full value.

Sometimes the bandwidth is referred to the low and high terminal frequencies where the distortion fails to rise above a specified value, the DIN value being 1 per cent.

QUIESCENT CURRENT

The static current drawn by the valves or transistors of a push-pull class B output stage as determined by the biasing in the absence of drive signal. See also under Class B.

RADIO BREAKTHROUGH

The breakthrough of modulated radio signals into the channels of an audio amplifier owing to the presence of high-level radio signal fields. The effect is that the base/emitter junction of the low-level input transistor rectifies the signals picked up by the wiring or circuit components and the resulting audio is then handled by the amplifier in the ordinary way so that the radio programmes appear as a disconcerting background on the wanted source signal.

Since the pickup preamplifier has the lowest level input, this source is most troublesome, particularly when the volume control is well advanced. In many cases the trouble can be cured without affecting the amplifier's

response or unduly changing the input impedance (when the preamplifier is in an equalising feedback loop) by connecting a 2000pF capacitor between the base and emitter of the pickup input transistor, using the shortest possible connecting wires. Both channels of a stereo amplifier must be processed.

RIAA

Standing for Record Industry Association of America and refers to the recording characteristic of gramophone records. The pickup equalising is a reciprocal of this characteristic.

RISE-TIME

Time taken by an amplifier for its output to rise from 10 per cent to 90 per cent of its final value when a perfect step (i.e., one with zero rise-time) is applied to the input. The rise-time is a function of the bandwidth.

The upper frequency response of most hi-fi amplifiers, with the treble filter off, is generally sufficient to give a rise-time well in advance of that of most programme material. (see also under Attack and Transient Distortion).

ROLL-OFF

The rate of attenuation of a filter at the bass end (highpass) or treble end (low-pass). The crossover frequency is that frequency where the response is —3dB of the mid-frequency response. The roll-off of an amplifier is similarly defined but in terms of decrease in amplification.

RUMBLE

Infrasonic disturbance due to the electro-mechanical action of a record player.

SENSITIVITY

Input signal voltage (r.m.s.) at 1kHz required to produce maximum amplifier output when the volume control is at maximum and the tone controls and filters set for a 'flat' response and, in the case of stereo amplifiers, with the balance control at neutral.

SIGNAL/NOISE RATIO

Voltage or power output due to unwanted noise referred to the full power of an amplifier or to some other datum. A S/N ratio of 60dB implies that the unwanted noise is 10⁶ below the full power or 10³ below the full voltage of the amplifier. The noise may or may not include hum. This is sometimes made clear by the parameter being expressed such as 'hum and noise 60dB below full output'.

TRANSIENT

Signal component of fast rising leading side and usually of short duration.

TRANSIENT DISTORTION

Distortion of a transient signal component such as produced by resonance and lack of damping, the effect being overshoot or ringing following the fast rising leading side of the signal component.

WATTS RMS

A commonly incorrectly used term for continuous wave power (see under Output Power).

WEIGHTED

Often applies to the filtering of a disturbance signal (such as noise) to take account of its annoyance value rather than its absolute value. A weighted S/N ratio is

always better than that obtained from the same amplifier (etc.) when unweighted.

TUNERS

AMPLITUDE MODULATION (AM)

Where the audio (or other) information applied to a radio wave causes changes in amplitude. Used generally in the long, medium and short wavebands but is not a hi-fi system as currently employed.

ATTENUATOR

Network for reducing signal level. Sometimes necessary in the aerial circuit of a tuner to avoid overloading by strong, local signals.

AUTOMATIC FREQUENCY CORRECTION (AFC)

Often termed automatic frequency control and is a circuit arrangement used in f.m. tuners to correct automatically for any initial, long or medium term tuning error. Currently based on a capacitor diode in the oscillator tuning whose reverse bias is adjusted by a potential delivered by the f.m. detector when the tuning is in error. In some applications it is desirable for the correction to be switchable to avoid the tuning pulling on to a strong station when it is required to receive an adjacent weaker one.

AUTOMATIC GAIN CONTROL (AGC)

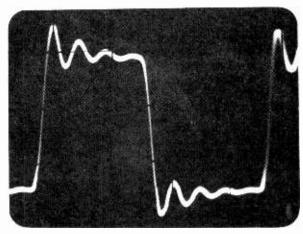
Circuit arrangement which yields a gain reducing bias with increase in signal strength as a method of keeping the output of the tuner constant irrespective of the strength of the received signal.

BALUN

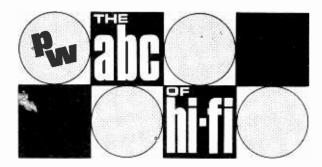
Usually a transformer designed to accept 75Ω unbalanced input (coaxial cable) and deliver the signal at 300Ω balanced (twin feeder). Usable in the converse sense, and sometimes necessary for matching a tuner with 300Ω balanced aerial terminals to 75Ω coaxial feeder.

CAPTURE EFFECT

Effect characteristic of the f.m. system whereby the stronger of two signals, even when in the same channel, 'captures' the tuner so that it responds mostly to this one, pushing the weaker unwanted one well into the background.



Transient Distortion. Example of transient distortion where the fast leading edge of a squarewave initiates ringing.



CAPTURE RATIO

IHF term which refers to the power ratio of two signals in the same channel required to keep the signal/interference ratio to a value of 30dB referred to 100 per cent modulation and 1mV input signal level. The ratio of the powers of the two input signals is expressed in decibels, the smaller the dB number the better the capture ratio. Top-flight tuners have a value as low as 1dB.

DECODER

Device used in stereo tuners for yielding the left and right channel signals from the multiplex signal fed in from the f.m. detector.

DISTORTION

Harmonic distortion produced by tuners is generally given at 100 per cent or 30 per cent modulation and 1kHz. F.M. detector non-linearity may cause the distortion to rise with increasing modulation percentage. Owing to the attenuation of the higher-order harmonics, the THD may be less from tuners with the American 75 μ S de-emphasis than from those based on the British (and European) 50 μ S de-emphasis. Distortion on stereo is also usually higher than on mono.

FIELD EFFECT TRANSISTOR (FET)

Transistor whereby the current is controlled by electrostatic means. These devices have greater overload immunity than bipolar transistors and a square-law characteristic (minimising third harmonic components) and for these reasons are favoured by many designers for use in the r.f. amplifier and mixer stages.

FREQUENCY MODULATION (FM)

Where the audio (or other) information applied to a radio wave (i.e., carrier wave) causes changes in the instantaneous frequency of the wave at a rate determined by the modulation frequency and by an amount determined by its magnitude. 100 per cent f.m. corresponds to $\pm 75 \text{kHz}$ deviation on the broadcasting system. F.M. is a hi-fi system of broadcasting also allowing the addition (in a subchannel) of stereo information. See under Decoder.

IMAGE REJECTION

This refers to the rejection by the tuner of a signal at the image (second channel) frequency, corresponding to the tuned (real) frequency **plus** twice the intermediate-frequency when the local oscillator is working above the signal frequency or **minus** twice the intermediate-frequency when the local oscillator is working below the signal frequency.

The IHF image rejection ratio refers to the ratio (in decibels) of the signal required for a 30dB S/N ratio to that required for the same ratio but at the image frequency. Front-end selectivity increases the ratio.

INTERMEDIATE FREQUENCY (IF)

The fixed frequency resulting from heterodyning (i.e., 'beating' or modulating to develop the sum or difference frequency signal) the incoming signal with a signal from the local oscillator. The i.f. used in f.m. tuners is commonly 10 7MHz and in am. tuners 470kHz. The i.f. signal is amplified in the i.f. channel and it is here where most of the selectivity is introduced by tuned bandpass couplings and/or crystal or ceramic filters.

INTERMODULATION

This can result in the tuner front-end owing to the application of two or more powerful signals. In the group of the three f.m. stations (Radios 2, 3 and 4), third-order intermodulation can put interference on Radio 3. The solution lies in reducing the strength of the signals by attenuation.

MULTIPATH DISTORTION/RECEPTION

Owing to reflection of the v.h.f. f.m. signal by large buildings, hills, etc., the tuner sometimes receives not only the direct signal but also a reflected signal slightly later due to the greater path distance travelled. This results in high-order harmonic distortion and impairment to the stereo separation and quality.

MULTIPLEX (MPX)

The name sometimes given to the stereo system of broadcasting and the complex signal delivered by the f.m. detector for application to the stereo decoder.

PILOT TONE

The signal included with the stereo information, and applied in a subchannel on the f.m. signal, which is used to reform the suppressed subcarrier and synchronise the switching of the detectors in the stereo decoder. Pilot tone frequency is 19kHz, and doubling this (38kHz) gives the subcarrier frequency. The pilot tone also activates the stereo indicator light switching.

QUIETENING (MUTING)

A muting circuit which 'quietens' the audio channel when tuning between f.m. stations and which is 'cancelled' when the incoming signal is of a pre-determined (preset) level.

QUIETING

The amount (expressed in decibels) by which the noise output of an f.m. tuner is reduced when the input signal is at a given level (see also under S/N Ratio and Sensitivity).

SELECTIVITY

The ability of a tuner to select the wanted frequency without interference from signals at adjacent frequencies. Front-end selectivity enhances the image (and i.f.) rejection ratio, while good i.f. channel selectivity prevents interference from unwanted stations at frequencies near to the tuned frequency.

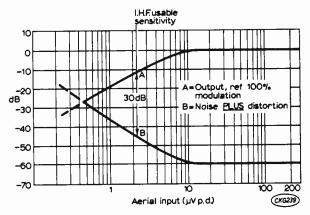
Good i.f. selectivity is required by f.m. tuners to minimise 'birdies' interference on stereo reception, coupled with low-pass filtering (critical frequency about 53kHz) between the f.m. detector and stereo decoder.

The IHF selectivity, given in many tuner specs., refers to the ratio of the level of a signal in the alternate channel (400kHz away from the required station) to the level of the wanted signal for a signal/interference ratio of 30dB when the wanted signal is applied at a level of $100\mu V$. The signal ratio is given in decibels, and the larger the number the

better the selectivity. Top-flight tuners with ceramic or crystal i.f. filters can have an IHF selectivity better than 50dB.

SENSITIVITY

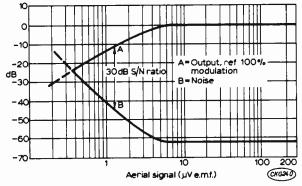
in general refers to the aerial input signal required for a given S/N ratio or quieting ratio (see S/N Ratio). The IHF usable sensitivity, frequently given in tuner specs., is the least aerial signal required across the tuner aerial terminals (p.d.) for a signal/noise plus distortion ratio of 30dB. Instead of the modulation being switched off it is notched out by a steep filter so that only the distortion and noise remains. The test is made with 100 per cent modulation.



Sensitivity. Graphical representation of IHF usable sensitivity.

SIGNAL/NOISE (S/N) RATIO

The ratio of the output signal to the noise at any given level of aerial signal. Signal modulation is either 30 per cent or 100, the latter giving an improvement over the former of about 9dB. A 0dB datum is established on the modulation, which is then switched off and the remaining noise measured. See also under Quieting and Sensitivity.



Signal/Noise Ratio. Curve A is tuner output referred to 0dB and 100% modulation. Curve B is noise output without modulation. Max. S/N ratio about 63dB while input of 1-4µV results in S/N ratio of 30dB.

SPURIOUS RESPONSE REJECTION

The ability of an f.m. tuner to reject spurious signals falling outside the tuned frequency or the immunity of the tuner itself to the production of spurious signals as the result of intermodulation, etc. (see under Intermodulation).

SUBCARRIER

The carrier used in stereo broadcasting to accommodate the subchannel stereo components. Frequency is 38kHz and is suppressed at the transmitter, leaving only the sidebands, but is reformed at the receiver for detection of the stereo components by doubling the synchronised 19kHz pilot tone (see under Pilot Tone).

TUNING INDICATOR

Often a meter which gives both an impression of the signal strength and the correct tuning point at maximum deflection. As maximum S/N ratio and least distortion occurs from an f.m. tuner only when the station is accurately tuned, a centre-zero meter might also be incorporated, and the tuning is carefully adjusted until this shows the correct condition of balance.

ANCILLARY

AMBIENCE

The term sometimes given to the 'colouration' of the signals due to the concert hall, etc. A kind of 'ambience' from stereo signals can sometimes be obtained by feeding the difference signals (appearing across the two 'live' terminals of the left and right channels) to a speaker sited behind the listener. More advanced matrices are necessary for full effect.

ANTI-SKATING BIAS

A bias force applied to a pivoted pickup arm to counteract the inward force (towards the centre of the record) resulting from the drag of the stylus in the groove and the offset angle of the head.

AUXILIARY BASS RADIATOR

A parastic (non-electrically-driven) unit resembling a bass speaker unit located in a loudspeaker enclosure, as if it were an ordinary unit, to increase the movement of alr and hence enhance the bass performance at a given enclosure volume.

AZIMUTH

Referring to the vertical setting (alignment) of the head in a tape recorder.

CCIR

Short for International Radio Consultative Committee, responsible for certain audio standards and tape play characteristics, among other matters.

CERAMIC PICKUP

A pickup whose generator system is based on piezo electricity produced by the stressing of natural and man-polarised crystals.

COMPLIANCE

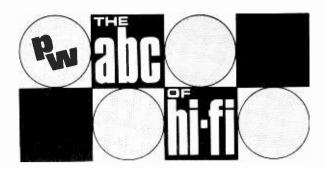
Opposite to stiffness. A mechanical unit whose electrical analog is capacitance. In gramophone pickups one 'compliance unit' is taken as 10⁻⁶ cm/dyne.

CONE BREAKUP

The inability of a speaker cone to work as a piston at high frequencies, the effect being that the cone is not under the complete control of the speech coil, certain parts of it moving in opposition to other parts like a 'rippled rope'. Responsible for uneven frequency response.

DE-EMPHASIS

A method of signal equalisation where pre-emphasis is used during recording or transmission. Such as the pre-emphasis of the treble modulation on f.m. and the equalising de-emphasis at the tuner. Used essentially as a noise reducing artifice.



DOLBY SYSTEM

A noise reducing system whereby the recorded signal is 'compressed' and the replay signal 'expanded'. Originally used professionally, but a simplified version for domestic applications (for tape recorders, especially those based on cassettes) has recently been developed.

DROP-OUT

Short pause in tape replay due to bad tape coating.

DUAL CONE

Speaker unit containing a main cone for bass and middle frequencies and a smaller, stiffer inner cone for treble frequencies, sometimes called a 'full range' speaker unit.

DYNE

Unit of force equal to 1/980 gram at latitude 45° at sea level. Used in pickup mechanics especially.

ELECTROMOTIVE FORCE (EMF)

Electrical pressure at source. Not to be confused with potential difference (p.d.) which is the voltage developed across a resistance or impedance due to current flowing through it. Both are measured in volts.

FLUTTER

Cyclic speed variation of tape recorder (e.g. above about 10Hz) producing spurious signal related to the frequency.

HEADSHELL

The end of a pickup arm where the cartridge fits. Sometimes bonded to the arm, though often detachable.

INERTIA

Resulting from mass and inhibiting change in velocity. Important in pickup mechanics.

INFINITE BAFFLE SPEAKER SYSTEM

A speaker system where the bass driver is located in an almost airtight enclosure.

LABYRINTH

A type of speaker enclosure where a maze-like construction extends the air column. Resonances are tamed by heavy damping material.

MASS

The bulk of matter though not necessarily equal to its weight. A mechanical unit whose electrical analog is inductance.

OMNIDIRECTIONAL

Having no particular direction of maximum emission or sensitivity. Omnidirectional speaker is one which radiates theoretically equally in all directions.

PARALLEL TRACKING ARM

Pickup system which allows the cartridge to track on the true radius of the record, as the recording was made, thereby minimising lateral tracking error.

PINCH EFFECT

Slight variation in the width of the groove resulting from the mechanics of recording which cause the pickup stylus to vibrate vertically at twice the recorded frequency, giving second harmonic distortion.

PLAYING WEIGHT

The downward force required on the pickup stylus to keep it in the groove and to counter the mechanical reactions of replay.

PRE-DISTORTION

A scheme sometimes adopted during recording to nullify the effect of tracing distortion on replay. The distortion deliberately introduced is the reciprocal of that produced on replay.

PRE-EMPHASIS

Deliberate boosting of high frequency to allow attenuation on replay (or reception in the case of f.m.) as a noise reducing artifice. See also under De-Emphasis.

QUADRAPHONY

A scheme of extended 'stereo' whereby ambient and dimensional information is fed directly or via a matrix to a set of four speaker systems suitably orientated in the listening room. Two main systems are currently under development. One based on four discrete channels and the other on two channels into which are matrixed the information of the other two channels. In the latter system a reciprocal matrix at the reproducing end delivers a version of the original four channels for application to separate amplifiers and speaker systems.

REFLEX SPEAKER

This differs from an infinite baffle enclosure in that an air vent is incorporated which is tuned by its dimensions for the best bass performance in conjunction with the bass driver and the volume of the enclosure. The auxiliary bass radiator is a development of this technique.

TIP MASS

The effective mass at the tip of the stylus of a pickup cartridge. Modern techniques have reduced this mass towards 1 milligram or below.

TRACKABILITY

How well the pickup system will track high amplitude and velocity modulation at a given tracking weight.

UNIPIVOT PICKUP ARM

An arm which secures both the lateral and vertical movements from a single highly engineered pivot. Difficult to engineer, but characteristics (i.e., very low bearing friction) desirable when well engineered.

VELOCITY PICKUP

Magnetic pickup whose output increases with increase of recorded velocity. This differs from the ceramic pickup whose output is a function of amplitude or deflection of the stylus.

wow

The complement of flutter, but at frequencies below about 10Hz.

ICSTABILISED POWER SUPPLIES

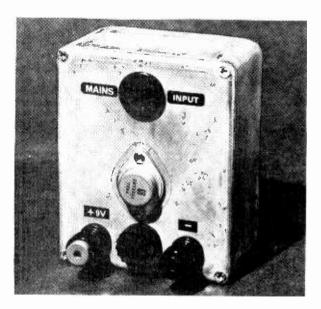
D.J.SILVESTER

HE design and construction of a compact high stability power supply results in a number of difficulties when the use of a 4½in. x 3½in. x 2in diecast aluminium box is contemplated. This particular supply unit has obvious uses as a 9V battery replacement, and since output current limiting can easily be added, the supply unit can be used in radio repair work where a current overload would otherwise cause irreparable damage. There is a limitation to the size of the mains transformer due to the smallness of the box, resulting in large variations of unregulated d.c. voltage, a difficulty which can be overcome.

The simple transistor-assisted zener circuit proved to be unstable because of these large voltage variations and the usual multi-transistor feedback circuits needed to give a stable output required more space than was available. The next step was to examine the possibility of using an integrated circuit as the feedback amplifier, and after this was done successfully a second and more powerful supply was built.

PRINCIPLE OF OPERATION

The block diagram of Fig. 1 shows the basic circuit of the two power supplies. The principle of negative feedback requires that part of the output voltage be compared with a reference voltage, and the output regulator drive adjusted so that the two voltages become equal. The amount of correction applied to the output regulator is controlled by the gain of the



Front view of the completed 9V supply.

amplifier. It can be shown that the stability of the regulated output depends upon:—

1. The gain of the amplifier.

- 2. The variations of the amplifier output with changes in the unregulated d.c. supply.
- 3. The stability of the reference voltage.

The integrated circuit has a very high gain (100,000 times) and has been designed so that a change of 1V in the unregulated supply will cause only a $30\mu V$ change in the output. The first two

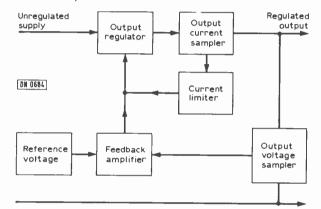
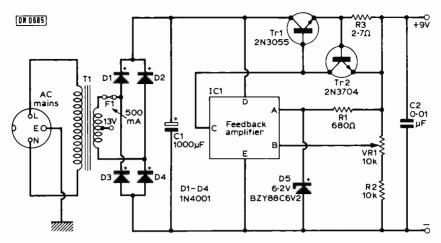


Fig. 1: Block diagram showing the various feedback paths of the stabilised supply.

sources of design difficulty are eliminated in practice if an i.c. is used as the feedback amplifier. The third source of instability due to the internal resistance of the zener diode can be overcome by providing a stable bias current from the regulated output, and by choosing a diode with a low internal resistance.

<u>FIXED</u> 9 volts supply

The final circuit used for the smaller supply is shown in Fig. 2, the few changes required to give a higher output power will be discussed later. It is possible to use either a μ A709C or μ A741C integrated circuit as the feedback amplifier, and as five possible package alternatives exist, this component has been represented by the box with connections A to E. These alternatives, as well as the three extra frequency stabilising components required with the μ A709C are shown in Fig. 3. It is preferable to use the μ A741C integrated circuit since this device will be undamaged should the inputs or



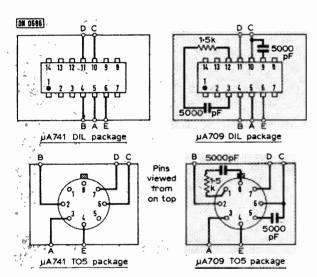
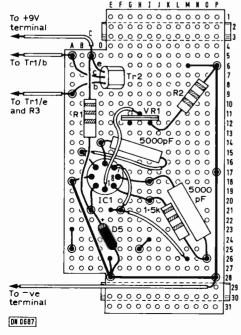


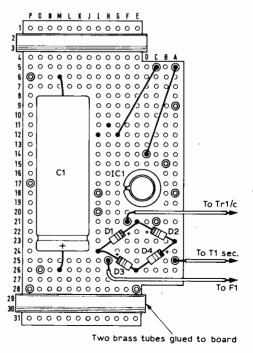
Fig. 2: Above, complete circuit of the fixed 9V supply. Fig. 3: left, connections of the various forms of the integrated circuit that can be used. Fig. 4: below, shows the veroboard layout for the smaller components. The 5000pF capacitors and 1·5κΩ resistor (see text) were fitted on the prototype.

★ components list

Tr1 2N3055 Tr2 2N3704
D1-4 IN4001 or 1A bridge rectifier
D5 BZY88C6V2 (6·2V 400mW)
IC1 µA741C (µA709C may be used, see text)
T1 230V/12-13V 500mA

Red and black insulated terminals. Die cast box, 4 x 3 x 2 in. Mains input plug and socket. Fuse 500mA and holder. Veroboard 3 x 1 in. 0·1in. matrix. insulating kit for 2N3055.



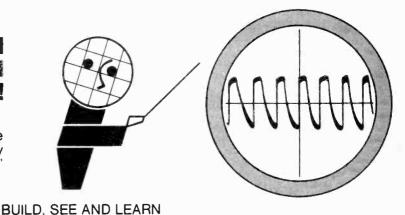


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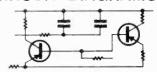


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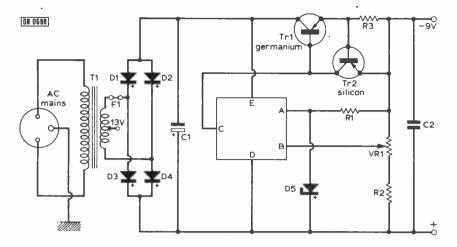


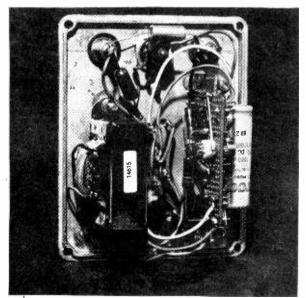
Fig. 5: Variation of circuit shown in Fig. 2 to permit the use of p.n.p. transistors such as may be found in an existing unstabilised supply.

output be shorted to either chassis or the unregulated d.c. supply.

The current limiting device consisting of R3 and Tr2 can be omitted if this facility is not required. The short circuit current with the components chosen is 250mA. The potentiometer VR1 is used in place of a single resistor in the output sampler, so that the output voltage may be set accurately to 9V when' construction has been completed. It was not necessary to reduce the sensitivity of this control by using a three resistor potential divider.

CONSTRUCTION

Very little information is required on the layout of the circuit components, which, with the exception of Trl', R3 and the transformer, can be assembled on a very small piece of veroboard (Fig. 4). One idea which greatly helps the final wiring is that all the components and board are mounted on the lid of the box. A neater appearance results if the larger components and the circuit board are stuck to the lid with epoxy resin adhesive.



Inside the fixed voltage supply unit. The transformer is fixed to the lid with epoxy resin as are short lengths of studding to locate the circuit board.

The power transistor uses the box as a heat sink but it must be electrically isolated with washers. The wiring should be thoroughly checked before switching on, after which the output can be set to 9V by adjusting VR1. A 47Ω resistor connected across the output should not cause any change in the output voltage.

VARIATIONS

It is possible to alter an existing power supply having germanium p.n.p. transistors, by using a variation of the original circuit, shown in Fig. 5. The unregulated d.c., supply must not exceed 36V or the i.c. will be damaged. If a different value of output voltage is required the values of R1, and VR1 and R3 must be recalculated.

If only a small current is required from the regulated output the power transistor can be replaced by a BFY51 or removed completely if an output of less than 10mA is required.

VARIABLE voltage supply

After the successful construction of the 9V fixed supply it was decided that a much more versatile supply could be constructed using the same basic circuitry. This new supply would be capable of powering small hi-fi amplifiers, and it was decided that in order to increase its usefulness the supply should be able to deliver a constant voltage in the range 10 to 25V at a maximum current of 1A. A few small design changes must be made to the circuit of the fixed supply, but one difficulty, the large variations in the unregulated d.c. supply, is to some extent overcome. The final circuit of the second supply is shown in Fig. 6.

The transformer used gave between 20 and 23V a.c. to the bridge rectifier consisting of four 3A 50V diodes, D1 to D4. These diodes produce a d.c. voltage of between 30 and 36V across the capacitor C1, the value of this component being increased to $5000\mu F$ by using two separate $2500\mu F$ electrolytics in parallel.

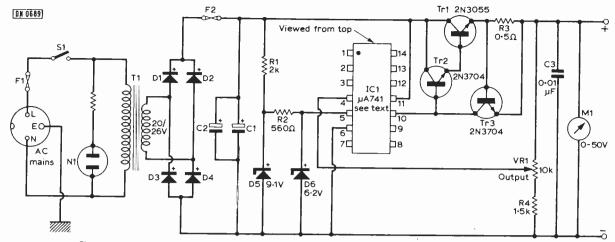


Fig. 6: Circuit of the variable voltage power supply which includes a meter to monitor the actual output voltage.

It should be noted that on no account, or under any load conditions, must the voltage across C1 exceed 36V or the i.c. will be destroyed.

OUTPUT CURRENT REGULATOR

The first transistor used for the current output stage is again a 2N3055 being the cheapest of a number of alternatives. The current gain of this transistor is about 20, which means that at maximum output current this transistor will require a base current of 50mA which is greater than the i.c. can supply. The final output current stage consists of a darlington pair Tr1, Tr2 using the 2N3055 and 2N3704/BFY51 transistors. It is calculated that the maximum current demand on the i.c. is now reduced to below 1mA. The value of the limited output current is 1·2A, obtained by reducing the value of R3 to half an ohm.

REFERENCE VOLTAGE

The reference voltage is derived from a zener diode D6, and as the stability of this component directly affects the stability of the output voltage

some method of supplying D6 with a constant current must be found. The zener cannot be supplied directly from the regulated output so the most obvious alternative is to provide a stable d.c. voltage from a second zener diode, D5, biased from the unregulated supply. The amplifier chosen is the μ A741C type used in the fixed supply but any of the alternatives given in Fig. 3 may be used if required.

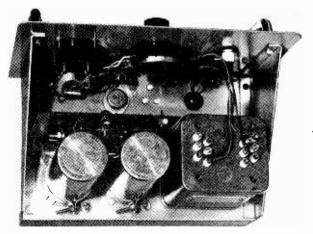
OUTPUT VOLTAGE SAMPLER

This part of the circuit consists of a resistor chain across the regulated output. The minimum output voltage occurs when the inverting input (lettered A in Fig. 3) is connected to the output rail, the output voltage being equal to the zener voltage of D6. The maximum output voltage is determined by the values of VR1 and R4. The value of R4 is calculated to be $1\cdot 3k\Omega$ for an output of 30V. The preferred value nearest to this, $1\cdot 5k\Omega$, produced a drop in the maximum output voltage. There is a rapid increase in voltage with rotation of the potentiometer at the high voltage end, and although this causes some slight inconvenience it was not considered necessary to use separate resistor chains.

-continued on page 642



Front view of the larger unit contained in a standard instrument case.



The smaller components are mounted on a vertical board shown above the large capacitors C1 and C2.

LETTERS.

Only knocks

I would like to comment on the correspondence regarding v.h.f. radio in your October issue.

Mr. James W. Robson's technical explanation for v.h.f. not having caught on is ingenious, but I fear, nonsensical. When f.m. was introduced in this country, and for several years after, only mains driven equipment was available for its reception. In those years, given suitable incentives, a large proportion of listeners could have been persuaded to buy f.m. radios. As it was, shackled by a monopoly broadcasting service, v.h.f. never stood a chance.

In the preface to "Radio & Television Servicing, 1955-56", a purely technical publication, the editors wrote: "... The outstanding feature of this period has undoubtedly been the inclusion, in the ranges of most manufacturers, of combined a.m./f.m. receivers for the reception of the BBC very-high-frequency (Band 2) transmissions . . ." In a subsequent edition it was stated that sales resistance to v.h.f. only sets was strong, and that the absence of Luxembourg on this band was partly to blame. And that is the situation in a nutshell. If only the opportunity had been taken at the time of the new channels becoming available to introduce independant radio stations, listeners would have clamoured for sets to receive them. Few saw the point of spending money to obtain carbon copies of existing programmes. A parallel may be drawn with the introduction of independent television on Band 3. In the two years following the opening of the Lichfield transmitter I supplied and installed upwards of 1,000 Band 3 converters. In the first two years of the BBC-2 625-line service I had orders for iust two conversion kits. Even to this day I find customers with facilities for v.h.f. radio and 625 line television who have no intention of using them. What a monu-) ment to wasted opportunity!

One might have thought that the instant and overwhelming success of the so-called "Pirate" radio stations in 1965 would be construed as a clear indication that something was seriouly wrong with broadcasting in this country. Attempts to legalise the pirates were frustrated by the BBC, who solemnly announced that it was quite impossible to find space for any more stations on the medium wave band. No one appears to have though of offering the largely unused v.h.f. band. Now the BBC has done a complete about turn by foisting its local radio stations onto medium waves. Nothing could be a greater admission of failure on the part of monopoly broadcasting to attract listeners to v.h.f.

promised commercial The stations may inject some life into v.h.f. reception, if only because all will be capable of transmitting in stereo, but by and large I am compelled to admit that I am far from sanguine. Even Lord Hill, contemplating the prospect of medium-wave reception in the future, is quoted as saying, "... and may the Lord have mercy on you after dark!" Unfortunately, f.m. broadcasting missed the boat 17 years ago, and opportunity, as they say, only knocks once. - Charles Miller (Uttoxeter).

Local radio

No doubt many readers will have noticed the additional interference caused in the m.w. band by the opening of local radio station m.w. transmitters recently, by the BBC. These transmissions, at night, can scarcely be described as local, with a range of over 200 miles. In my location, Radio London is very powerful, and has been received at night in Holland, France and Northern England. This rather defeats the BBC's good original object of a local radio service. Could not the BBC shut down the 206 metres Radio London m.w. outlet, and other outlets, at night, when local radio simply relays programmes carried by Radios One and Two. The v.h.f. outlets could of course remain, as they have done previously. This step would reduce the amount of interference with foreign services, and would bring BBC local radio in line with the proposed restrictions to be imposed upon local commercial radio stations.—John Manners (Middlesex).

Quadraphonics

I am interested in building myself a Quadraphonic Hi Fi system.

The design of the preamplifiers and power amplifiers is relatively straightforward, but as yet I have been unable to find any details on how to design the four channel matrix decoder.

An article in your magazine on either the principle of operation or a practical circuit would be of considerable value to myself and I am sure to other people.—

P. M. Cunnington (Essex).

Your letter has been forwarded to me by the Editor and I must comment, with a copy to the Editor, in the hope he may publish, for your letter raises one or two interesting points.

First, before we could contemplate publishing the circuit for a decoder, we must know something of the encoding. Quadraphonics is very much in a state of flux.

In April, there was a show at Osaka, and about half of the 38 exhibitors had some sort of four-channel exhibition. Few were convincing. Most annoying was the fact that few were mutually compatible—nor even tried to be!

There are, basically, three systems of stereo recording: the CD-4 of Matsushita and JVC-Nivico, with American backing from RCA; the SQ, with CBS and Sony throwing in their weight, to which EMI have now added their backing in this country. And last is what might be called a 'Regular Matrix' system, like the Sansui QS, and several related alternatives.

There is no 'universal' decoder. and until the vast commercial battle is won and there is a 'standard' method of encoding, we should play it very close to the chest. I would advise you, Mr. Cunnington, to forgo the pleasure of anything but purely discreet quadraphonics until your way ahead is sufficiently clear. In any case, you can go ahead with preamps and power amps-and, I hope, separated power supplieswith the knowledge that whatever decoder is appropriate to the software available to us (mostly SQ at present, will soon be made available to the constructor .-H. W. Hellyer (Bristol).

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	K.E.F.B.200		£15:00 A
	K.E.F.B.110		£14 00 A
	K.E.F.N.139 MKII		£18-50
	Wharfedale Unit 3		£18:00 A
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	Wharfedale Unit 5		£35-50 D
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	Peerless 10-2 Kit		£16:00 A
	Peerless 3-25 Kit		£30 · 00 C
	Peerless 20-2 Kit		£20.00 B
	Peerless 20-3 Kit		£31 · 00 B
	Peerless 3-15 Kit		£21-00 B
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Phillips El 2202	£15.00 A	Standard Version
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Carriage & Packing A	k-50p, B-£1.00	Garrard SL72B
CARTRIDGES		Garrard SL65B
Audio Technica AT 6	56 £3·75	Garrard 2025 TC with
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Amstrad 900C	£3-45	Garrard Zero 100S
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A.D.C. 10/E Mark 4	£20.50	B.S.R. MP60 TPD1 B.S.R. MP60 TPD2 B.S.R. MP60 TPD2 B.S.R. HT70 B.S.R. HT70 P-C
A.D.C. 26	£44.00	B.S.R. MP60 TPD2
A.D.C. 25	£63 00	B.S.R. HT70
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Goldring G800 Goldring G800/E Goldring G800 SE	£14-25	Hydraulic REF with 9" or 10
Shure V.15 Type 2 Im	ull range available£1 · 45£3 · 20£5 · 00£7 · 00£14 · 25 pr£24 · 50£10 · 25£8 · 85	I Goldring GL/85/P with IId
Shure M75ED	£10 25	Goldring GL 75
Goldring G850 Goldring G800 Goldring G800/E Goldring G800 SE Shure V.15 Type 2 Im Shure M75EJ Type 2 Shure M75EJ Type 2 Shure M75E J Type 2	£8⋅95	Goldring GL 75
Shure M75-6	£7·25	
Shure M75G	£7·95	Goldring GL 72
Shure M55/E	£5 65	Goldring GL 12/P
Shure M44/E	£7 - 25 £7 - 95 £5 - 65 £4 - 90	Goldring GL 72
Shure M44C.5.7.	£4-55	Goldring G101/p-c
Shure M75-6 Shure M75-G Shure M75G Shure M55/E Shure M44/E Shure M44C.5.7. Shure M3D	£3 94	Goldring G101/p-c
COMBINATION UP	NITS/COMPACT	Thorens TD 150AB
STSTEMS		Thorens TD 150A
	KS . £63 00	
Dhillian CE cool . C		Thorens TX 11
Philips GF 808 Inc Sp	ks £77 00	Thorens TX 11 Thorens TD 150
Philips GF 808 Inc Sp Philips GF 805 Inc Sp	ks £63 00 ks £77 00 ks £59 00	Thorens TX 11 Thorens TD 150 Thorens TD 125 mk II
Philips GF 808 Inc Sp Philips GF 805 Inc Sp Philips GF 980 Inc S Philips BH 811 Inc Sp	ks £77 00 ks £59 00 pks £117 00	Thorens TX 11 Thorens TD 150 Thorens TD 125 mk II Thorens TD 125 AB
Philips GF 808 Inc Sp Philips GF 805 Inc Sp Philips GF 980 Inc S Philips GF 980 Inc S Philips BH 814 Inc Sp Philips BH 814 Inc Sp	ks £77 00 ks £59 00 pks £117 00 ks £95 00	Goldring G101/p-c Goldring G99 Thorens TD 150AB Thorens TD 150A Thorens TX 11 Thorens TD 150 Thorens TD 150 Thorens TD 125 mk II Thorens TD 125AB MK II
Philips GF 836 Inc Sp Philips GF 808 Inc Sp Philips GF 805 Inc Sp Philips GF 980 Inc S Philips RH 811 Inc Sp Philips RH 814 Inc Sp Philips RH 814 Inc Sp	ks £77 00 ks £59 00 pks £117 00 ks £95 00 ks £109 00 ke £54 00	Thorens TX 25
Philips RF 836 Inc Sp	ks £72 88	Thorens TX 25
Philips RF 836 Inc Sp	ks £72 88	Thorens TX 25
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Philips RF 836 Inc Sp Philips RH 802 Philips RH 891	ks £72 88	Thorens TX 25 Pioneer PL 12AC Pioneer PLA 25 Philips GA 202 A with Cart.
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THE r.f. signal generator forming the subject of this article was designed for the alignment of superhet receivers as well as for general purpose usage. The cost should be low as the coils are of the Denco plug-in type and do not form a permanent part of the circuit. This has additional advantages in that range switching is not necessary and all r.f. leads and connections can be kept very short; construction is also simplified greatly by using this method. The coverage of the generator depends entirely on the reader. Two coils will cover the medium waves, long waves and the i.f. of 465kHz but the prototype also worked on the short waves up to at least 10MHz; this however is not so vital as the harmonics of the lower ranges will get up to this frequency and way beyond. The circuit used has proved very reliable and gives a good output.

THE CIRCUIT

The complete circuit is given in Fig. 1 and may

be considered in two distinct parts, the r.f. oscillator and the a.f. modulator. The r.f. side is very simple and comprises a tuned circuit (VC1 and La) with base coupling (Lb) with a feedback winding (Lc). Base bias for the transistor is provided through R1 while C1 decouples the supply line with the take off point being C2 from the collector of Trl. A BC169C is used, mainly because of the lead configuration; this is a plastic encapsulated version of the BC109.

HALVOR MOORSHEAD

The supply for this stage is provided through a dropping resistor R2.

The a.f. section comprises Tr2 and associated components. This is a Hartley configuration using a transistor output transformer type LT700. This is a completely conventional arrangement, the primary of the transformer is tuned by C4 with C5 providing the feedback coupling and R3 the base bias to start the operation (this resistor may not be necessary as the feedback should be of sufficient level to maintain oscillation but it is as well to include it to make sure).

The output from the a.f. oscillator is at a fixed frequency-in the case of the prototype it is about 800Hz but the exact frequency will depend on the particular transformer and on component tolerances etc. The frequency can be increased by lowering the value of C4 or omitting it altogether, relying on the self-capacity of the windings to tune it. The frequency can be lowered by increasing the value of C4.

The waveform from the audio oscillator is far from a pure sine-wave but this is of no consequence and serves the desired purpose admirably.

FUNCTION SWITCH

Switch SW1 is the main function switch providing the on-off facility, r.f., modulated r.f. or a.f. at the output socket. In position A the battery is unconnected and so the circuit is off. In position B the

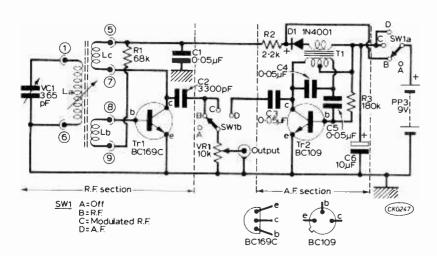


Fig. 1: The complete circuit of the R.F. Signal Generator. The required range is selected by plugging in the appro-priate coil. SW1a and SW1b are ganged to give the required output.

positive supply is connected between D1 and R2. In this mode the diode will block any current to the a.f. oscillator and this will be inoperative. Current, however, is allowed to reach the r.f. stage and SW1b connects the output level control to the output of

Trl via the d.c. blocking capacitors.

In position C, VR1 remains connected to the output of the r.f. section but SW1a applies current directly to the a.f. oscillator stage. In this mode D1 allows current to reach Trl but the positive supply is now connected via the secondary of the a.f. oscillator transformer. This creates a small, rapid change in the supply voltage in sympathy with the frequency of operation (800Hz in the case of the prototype) and so the output from Trl will also vary, modulating the signal. Normally it is not good practice to modulate an oscillator as this can cause both amplitude and frequency modulation but in practice some slight f.m. can be ignored. This arrangement gives a modulation depth of about 20 per cent.

In position D the positive supply is as for C but SW1b connects to the collector of Tr2 via the d.c. blocking capacitor C3 giving an audio output. If it is important that only pure a.f. is provided, the r.f. section can be made inoperative simply by unplugging the coil.

Capacitor C6 decouples the supply of the a.f. stage, the relatively low value is quite sufficient

because of the audio frequency used.

The output of the r.f. section is about 500mV and that of the a.f. section about 400mV; in both cases this is more than adequate for normal usage and both outputs connect via the level control VR1 before connecting to the coax output socket.

CONSTRUCTION

The coils used are the Denco plug-in types, these fit into a B9A valve-holder. The leads can be kept very short by building the r.f. section directly onto

* components list

Resistors R1 $68k\Omega$

180k Ω

VR1 10k Ω lin. pot. R2 2·2kΩ All fixed resistors \(\frac{1}{2} \text{W} \), 5 per cent tolerance

C1 0.05µF ceramic disc

C2 3300pF ceramic, polystyrene or silver mica

C3 0.05µF ceramic disc C4 0.05µF ceramic disc

C5 0.05μF ceramic disc

C6 10µF 12V electrolytic

(Ali 0·05μF can be 0·047μF) VC1 365pF tuning capacitor, Jackson Type 00

Semiconductors

Tr1 BC169C

D1 1N4001 or similar

Tr2 BC109

Miscellaneous

Colls: Denco transistor range, Blue, 1T, 2T, 3T of to suit; T1: Eagle transistor output transformer type LT700; SW1: 4-way, 2-pole rotary type; B9A valve holder; Coax output socket; Veroboard 0.15in. matrix 16 x 13 holes; PP3 battery and battery clip; Terry clip for holding battery; Aluminium angle; Plastic sheet for pointer; Knobs to suit; Metal chassis 5 x 4 x 24in.

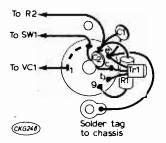


Fig. 2: The wiring of the r.f. section on the B9A valve base into which the coil is plugged.

the pins of the valveholder as shown in Fig. 2. To neaten the layout some changes have been made to the recommended Denco connections but the electrical result is the same. To bring the negative chassis connection to this part of the circuit a short wire runs to a solder tag fitted underneath one of the valveholder mounting screws.

Pin 4 of the valveholder does not connect to the coil and is only used as a convenient anchoring point

The layout and construction of the a.f. section is much less critical than that of the r.f. oscillator and can be built in any convenient form. In the prototype these components are mounted on a piece of Veroboard of standard 212in. width, 13 holes long. The layout of this section is shown in Fig. 3.

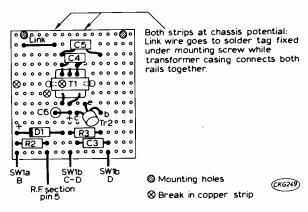


Fig. 3: The component layout of the a.f. section on Veroboard.

Holes should be drilled out to take the mounting screws and the lugs of the transformer. There are two strips on this board which are connected to chassis potential, these are linked together through the transformer casing, the lugs being bent inwards and soldered to the adjacent strip. One of these strips also connects to a link wire which runs to a solder tag fitted under one of the component board mounting screws.

There is no crowding of the components and construction should present no problems.

A few breaks are necessary in the conductor strip; one underneath the transformer is absolutely necessary, the other two will only be needed if the solder tag or the transformer lugs touch an adjacent

Four take-off points are required; these are shown in Fig. 3 and Fig. 4, the latter shows the wiring not dealt with in the other drawings. A stiff wire should

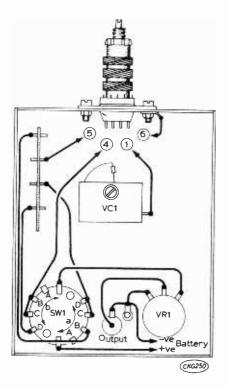


Fig. 4: The wiring between the controls and the two main operative sections.

be used from the fixed vanes of VC1 to pin 1 of the valveholder.

The rotary switch used is unlikely to look like that in Fig. 4, it is only shown in this way for clarity. The coax output socket used was of the insulated type and so a chassis connection has to be made to it; more conventional types will not need this.

It is very important that a good quality valveholder is used. Initially the author used rather a poor one and this later had to be replaced. The frequent plugging and unplugging of the coils demands a first rate quality component here.

The pointer can be made from any thin rigid clear plastic; the dimensions for this are given in Fig. 5a. A score mark should be made centrally on the underside, this can be inked in to make it clearer.

The component board is fixed to the chassis by

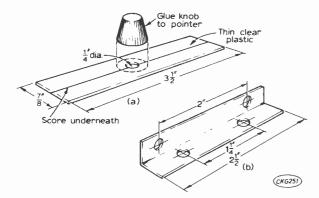
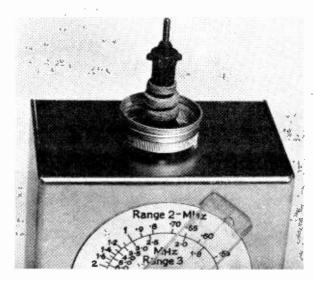


Fig. 5:(a) The construction of the frequency pointer.(b) The mounting bracket for the a.f. section component board.

means of an aluminium angle bracket as shown in Fig. 5b.



The screening can lid should be fixed on top of the valve base, ensuring that there is an electrical connection.

The main case is a $5\times4\times2^1$ 2in. chassis—this is a fairly common size and the drilling of the face is shown in Fig. 6. The only part of this which may cause problems is the mounting of VC1; the dimensions of these holes and their siting are rather critical. It is best to first drill a 1 2in. diameter hole, this will then accommodate the raised section around the ball race. The capacitor itself can then be used as a template for finding the exact locations of the two mounting screw holes. Very short screws must be used to mount this component, if they are too long they will foul the moving vanes of the capacitor.

The valveholder is mounted centrally on the top

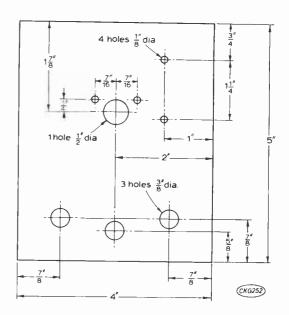
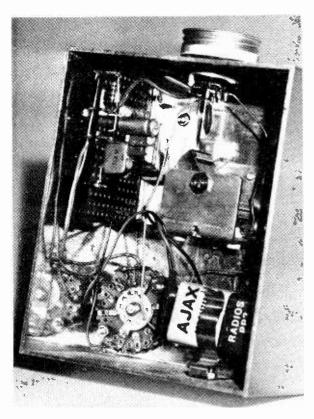


Fig. 6: The drilling details of the chassis face.



Rear view of the completed prototype.

of the case; no problems are likely to be associated with this and so dimensions are not given.

Battery mounting is often a problem; it has been solved in this case by using a Terry clip which grips the PP3 battery; this is fitted behind the level control.

To have control over the output level, the coil should be screened; if it is not it will radiate considerable power. The screening can supplied with the coils may be used for this. A 34in. diameter hole should be cut in the lid and this glued over the valve base at the same time ensuring that there is an electrical connection to chassis. A hole should also be drilled in the top of the can to provide access to the coil core adjusting screw.

CALIBRATION

Articles on r.f. signal generators frequently make a lot of calibration, this however need present no problems if you have available an all-band radio. Radio stations maintain very accurate frequency control and by beating the r.f. of our signal generator with their carriers, we can obtain very accurate calibration.

A blank calibration panel is given in Fig. 7. This can be cut out and stuck onto the face of the case or copied.

It is best to calibrate the long wave band (range 1) first as this will also give us our i.f. frequency. Assuming that all is well with the circuit and wiring, when the range 1 coil is plugged in and the function switch is set to MOD, the output should be tuneable over the long wave band and this will be heard on a radio. A wire should be connected to the inner of the coax socket and placed near the radio, the level

Easily received MW/LW Radio Stations that can be used for calibration. Extra points can be found by interpolation

	Range 1	Range 2			
Frequen	cy Station	Frequency	Station		
151 180 200 233 245	Deutschlandfunk Europe No. 1 BBC Radio 2 Luxembourg Denmark	529 602 647 701 746 800 908 1007 1106 1196 1295 1403 1502	Algeria France BBC Radio 3 Andorra Hilversum 1 W. Germany BBC Radio 4 Hilversum 2 AFN Munich VOA Munich BBC European Service France Warsaw (Often English in late evenings) BBC Radio Leicester		

should be turned down as low as possible to prevent pulling in the receiver.

The receiver should be tuned to the German Deutschlandfunk programme on 151kHz—this is usually well received throughout Europe. With the vanes of VC1 fully closed and the pointer lined up with the end of the scale the core of the coil should be adjusted till the a.f. note, combined with a beat note, is heard. When this is done, tune VC1 to a point where the vanes are almost completely open. If the receiver is a superhet with an i.f. of 465kHz (or 455kHz for many imported sets) another note will be heard. Unlike the 151kHz point this note will be heard from the receiver irrespective of the band or frequency (though not on v.h.f.) the receiver is tuned to. When this point is found it should be

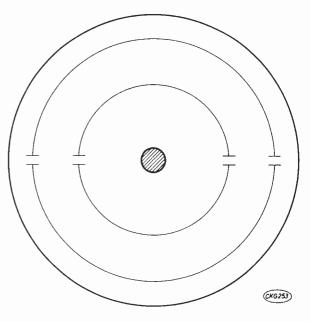


Fig. 7: This basic scale may be cut out and calibrated as described in the text.

carefully marked as this is the i.f. If possible, two positions should be marked, one for 455kHz and another for 465kHz.

Range 2 should be adjusted in a slightly different way. If we adjusted the core of the coil to operate on, say, 529kHz (Algeria), the high frequency end of the medium wave band would be badly cramped as this point would be reached before the vanes of VC1 reached their "extra spread" setting. For best results the core should be adjusted so that at a frequency of about 700kHz coincides with the vanes half open (pointer vertical). Andorra on 701kHz can be used for this. The range should then extend from 500kHz to over 2MHz, encompassing the complete medium wave band and the 1.6MHz i.f.

Range 3 should be adjusted so 1.6MHz coincides with the vanes fully closed. This range should then go up to about 6MHz.

The harmonics of range 3 can be used to give coverage of up to 30MHz though these are of less use and will be of lower level.

Various low level whistles may be heard on the wrong frequencies-these can be due to harmonics or receiver action but they should always be at a low level and there should never be any real doubt about determining the fundamental, it will be by far the strongest.

A table of useful stations for helping in the calibration are given; for a more complete list. reference books such as the "Guide to Broadcasting Station" or "World Radio TV Handbook" should be consulted.



Yorks, W., for Jan, and Feb. 1971 containing details of Stereo tape recorder construction.—M. Scott, 13 Upper Montagu St., London, W.H. 1RR.

"P. E. Feb. 1988 containing I.C. tape recorder.—D. Hollis, 15 Merlin Close, Coiller Row, Romford, Essex.

"Wanted P.W. October 1971 and April 1972. Postage and cost refunded.—K. Foster, 148 Malmesbury Road, Shirley, Southampton, SCI 5EY.

"Feb. and March Issues of 1971 containing the P.W. Stereo Tape Recorder.—D. J. Thomas, 16 Western Road, Urmston, Lancs. M31 3LF.

"April and May 1980 copies of P.W. and any information on 19 set Mk. III.—T. Izard, 41 Clumber Drive, Radiff-on-Tent, Notts.

"F.T. Feb. 1986.—J. S. Fairie, 5 Douglaston Crescent, Milngavie, Glasgow, Scotland.

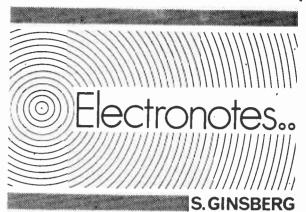
land. ...P.W. Jan. 1972 and Feb. 1972 issues.—C. W. Lo, 183 Lodge Avenue, Dagenham,

...P.W. Feb. to June 1967 and Aug. to Dec. 1970. P.E. Volumes 4, 5, 6 and Feb. 1971, All in very clean condition for binding.—E. Somerville, 2 Hillcrest Place, Killwinning.

All In very clean condition for binding.—E. Somerville, 2 Hillcrest Place, Killwinning. Aryshire, which applies the first place of the condition of the conditi

.W. Jan. 1972 and Feb. 1972.-B. A. Cotton, 15A, Thomas King House, Welling-

...P.W. Jan. 1972 and red. 1972.—D. A. Cotton, 1943. Historias hing house to const. Coventry.
...P.W. Aug. 1969 to September 1971 inclusive. Price please to.—P.O. Rox 341. Hastings, New Zealand (Airmail).
...P.W. Feb. 1972.—P. Parker, 31 Fair Vlew, Handsacre, Rugeley, Staffs, WS15 4DJ.
...April 1966 P.W. Including 191 Set Mods. Article.—A. C. Arthur, Pinkworthy Farm, Pyworthy, Holsworthy, Devon, EX22 64Q.



THE art of sonar is not quite as simple as it sounds. In radar, an electromagnetic wave is transmitted in bursts and the energy from

these which is reflected by surrounding objects is viewed on a cathode ray tube. Different objects cause different amounts of reflection and thus cause a pattern to be set up.

Sonar does the same sort of thing but with two major differences. First, it uses sound waves instead of radio waves. Secondly, it functions under water, the classic application being in ships.

Basically, sonar sounds simple until one considers it in greater depth-so to speak! A moving target generates acoustic energy, a form of noise, which can be detected. Thus no generated pulses would be needed in this passive mode. Perhaps, already I have stimulated anglers into thoughts of being able to actually "see" the big ones or shoals? Incidentally, professional systems do exist which can detect a single cod at 600ft!

In the case of sonar, the speed of data collection is limited mainly by the velocity of the propagation ofacoustic energy through the water. In sea water, this is about 1,500 metres/second. Although this may sound (sorry!) fast, compare it with electromagnetic energy which has a velocity of propagation of some 3×108 metres/second. Thus radar would take only 1/1,000th. of a second to receive a reflection from a target nearly 100 miles away. For sonar, it would take something in the order of four seconds for a target about 3,200 yards away-barely two miles.

Like its r.f. counterpart, sonar has problems with noise. The surface of the sea is in a continual state of variation due to the weather. Marine organisms cause noise, so does seabed turbulence.

At u.h.f., microwave energy in the earth's atmosphere travels in a straight line tending to follow the curvature of the terrain. Unhappily, sonar's acoustic energy is very reluctant to follow any straight line in sea water.

To complicate matters further, water temperature has an affect. In the sea, there is a temperature gradient. One "layer" of water is a little warmer than the next one down which in turn is a different temperature to the one immediately below it. It is not until nearly a mile down that an almost constant temperature is reached.

It is interesting that the velocity of acoustic energy varies dependant upon the temperature, pressure, salinity etc. Thus the acoustic waves are bent or diffracted (like light waves in water and glass) the amount of diffraction depending upon

—continued on page 641

practically wireless commentary by HE

HENRY

No. 93 Do you read me?

OME while ago, Henry had the uncomfortable pleasure working beside a pedantic colleague who corrected every casual remark as if invested with the guardianship of the English language.

A favourite dampener was: in answer to a query, "What does the meter say?" "The meter doesn't. You have to read it!"

Recent forays into the world instrumentation have convinced this traveller that we shall soon render Old Niggly out of date. We have lately been promised computer type voltmeters that will store a measurement and notify a comparison. A sort of digital poor cousin of the storage 'scope. It is but a small step to the meter that will shout: "Hey! Give over" when overloaded, or a prim: "Three-pointseven millivolts" at the appropriate moment. With, of course, a built-in recorder (opto-magnetic; I mean to invent it) so that Old Niggly cannot dispute the matter, though he is sure to contest the manner of its delivery.

Avid and persistent collectors of Henribilia will by now have gathered that the subject of digital read-out devices for the ordinary radio workshop occupies the upper layer of the author's



The meter that will shout "Hey! Give over!"

(for want of a better word) mind.

It is true. I have recently come into proud possession of a digital multimeter.

Quite apart from a phenomenal accuracy, better than the thickness of a conventional pointer on the lowest voltage or highest resistance scale, the Digivom has the advantage of push-button range selection, and a read-out large enough to satisfy the most myopic niggler.

Mind you, it is not all plain sailing. There are one or two disadvantages the brochure omitted to mention. One is the tendency the Digivom has of sending out its own identifying signal—a series of chopped pulses that render its use as a measuring instrument for delicate audio (and some radio) purposes, about as valid as a Bobby Fischer unpaid checkmate.

The darned thing radiates all through the workshop. And if you enclose it in a tin box, leaving only the display tubes glowing through their visor like the Red Knight's menace, it still manages to jigger the f.m. band on the next-door engineer's rush job.

At the moment, I have a Solartron rival which uses diode display devices and these radiate only a mild hash, a, presumably there is hope for we humble servicemen yet. Except that this one has a read-out much smaller and of a style which, file the numbers my friendly, neighbourhood bank manager scrawls along the bottom of my cheques, defies analysis except by logical devices.

It is fascinating to watch the digivoms trundling down through the voltages when one reads off the h.t. line. Almost like a Cape Kennedy transmission!

Henry was happy with his pushbuttons till Old Niggly sowed the seed of doubt in his mind. "Why not", he said, "dò away with the push-button rangè selection?" After all, if you can read-off an overload indication and produce



Throw one or two in the makers' faces.

a flashing light, and do a similar service for the wrong d.c. polarity, why not build in a trigger circuit fed from a sensing device so that the meter must always switch itself to a range appropriate to the voltage applied to its terminals?

Well, why not? I asked an electronics designer whose name you will have seen if you read the glossier audio journals or works on instrumentation. He said clamly: "It has been done," and quoted me one of his own patents.

Now Henry doesn't pretend to be a designer or an inventor, and his only claim to-innovatory fame is a range of hook-up switch-boxes. But, fired by the ambition engendered by necessity—and not a little niggled by chagrin at spending over a hundred quid on a digivom before discovering its drawbacks—he has sworn to throw one or two modifications in the makers' faces before the year is out. Watch this space for details. . . .

The modifications will not, true, confound Old Niggly by making the machine shout its readings to all and sundry. I cannot hope to incorporate storage facilities. But the least that can be done for an instrument that boasts only one current scale, and that the "basic one", is a switched range of external shunts. *That* should not be beyond Henry's abilities—when he gets five minutes to spare.

Where can I get exclusive units at lowest prices?"...



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RP-1000ST which has full record and playback lacilities. Automatic track change with manual over-ride

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REALISTIC 36 WATT STERED AMPLIFIER MODEL SA-500 0 A superb hi-li amplifier with all the leatures you've ever wanted - for under

February 1972 tures you've ever wanted - for under £46-00. Saving over £10-00 on the normal retail value. Up-to-the-minute stider controls for bass and treble £35 volums and balance controls. Headphons socket on front panel, Push-buttor input controls - magnetic phono (high/low) tuner, aux, mono, monitor Loudness push-button control for perfect sound at low output levels. Left and right push-button on/off switches for speakers. Noise filtering

and tape monitoring facilities. Two auxiliary AC outlets, frequency response 20-20,000 Hz \pm 1 db at full power. 15 watts rms per channel. Walnut cabinet with satin aluminium trims. Inputs: phono 2.5mV and 5mV RIAA: tuner/aux 250mV. Hum and noise: phono - 50db, tuner/aux - 65db. How's that for a specification! Size 14½ "wide, 3½" high, 10½° deep

OLSON RA-316 AM/FM/MPX STEREO TUNER This ROC Tuner is aspecially de

signed to match the Olson AM-395 Stereo Amplifier. In price and value, as well as it's good looking design! But of course it's also for use with any other amplifier. The RA-310 costs £10 00 less than the normal retail value, and yet it is a highly sophisticated unit, incorporating the latest solid state techniques. Operation is dirt life for supporter existing the latest solid state techniques. Operation is dirt life for supporter existing holding capability. You can connect this Turer to a stereo

Normal Price £50 00 amplifier to a tape deck or a tape recorder. And of course it covers all the stations in the AM and FM bands. FM. 87-108 MHz, AM. 525-1605 kHz. FM. Sensitivites: FM. 3pV. AM, 250pV. Stereo separation 30dB at 1kHz, image rejection 60dB. Size: 11-2, wide, 4" high, 7;" deep.

1 3 £12.45

REALISTIC SA-1008 6-WATT

Here's fabulous, exciting value in miniature! This high quality stereo amplifier measures only 9" wide × 3" high × 5% deep. And yet it has sepa rate ganged volume, balance and tone com thats. Plus speaker in/out, mana/stereo, phono

tuner and power on/off slide switches. The ends are oiled walrut, with matching in genamelled metal top. The front panel is satin aluminium and walnut-brown enamel. Frequency response is 50 to 10,000 Hz \pm 3d8. Output 3 walts r.m.s. per channel into 8 ohms. Inputs are 100mV for both phono and tune

EAGLE LC.05 STEREO MAGNETIC CARTRIDGE For fabulous reproduction at a very low price you'll find it hard to beat

0-7 mil diamond stylus. Out-put: 6mV per channel. Frequency range: 30-18,000 Hz. Channel balance: ±1-5d8. Channel separation: 20d8. Recommended stylus pressure: 2-4 grams: Compliance 9×10-6 cm/dyne ROC PRICE £4-75

EAGLE LC.07 STEREO 🖣 MDVING-MAGNET

Here's your opportunity to own a transcription castridge for the price of a ceramic! Is specially de signed to match top quality tons arms, and to get the very best from your hi-fi amplifier. 0.7 mil diamond stylus. Output: channel, Frequency range: 20-21,000 Hz Channel balance: = 1dB. Channel separation: 28dB. Compliance: 12 x 10-6 cm/dyne ROC PRICE C6 37

OLSON AM-395 40-WATT STEREO AMPLIFIER

An ideal unit for your new stereo separate system. It is more than £10 00 below the normal re ail price! Making the AM-195

one of Britain's best h-fi buys. It takes in signals from magnetic or ceramic pick-ups tuners (see Olson RA-310) and tape decks. And it's got out-puts for taping and for headphones. There are separate bass and treble controls, separate Left and Right channel volume controls. And a loudness switch for boosting the bass and treble notes when listening at low output levels. Frequency response: 20-20,000 Hz ± 3dB. Dutput: 20 watts r.m.s. per channel

into 8 ohms. Inputs: magnetic phono 3:0mV RIAA, crystal phono 100 eV; tape 160mV; tuner 160mV. Size 11 2 wide. 4 high, 72 dasp. The specification reads well - sounds even better!



A-3000 36-WATT SOLID STATE STEREO AMPLIFIER The A-3000 looks as good as it sounds! Giving you a big performance this superb audio amplifier has a full range of facilities on the front and rear panels. On the front all the controls you're ever likely to need plus a headphone socket. On the rear -signal inputs, speaker outputs and a line

luse for circuit protection.
Specifications: 18 watts rms per channel into 8 ohms. Frequency response 20-35,000 Hz (± 2db) Inputs Magnetic, Ceramic, Tuner, Tape, Aux. Tape Play. Size: 345mm × 300mm × 130mm.

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£15.70

DLSON AM-357 4-WATT STEREO AMPLIFIER 🝳 Here's marvellous value for someone just starting to set them-selves up in audio! At only £10 50, you get a fine amplifier in a scratch resistant metal cabinet, with a smart brushed aluminium front panel. It incorporates separate tone and volume controls for each channel. Inputs are provided for turntable (ceramic car tridge), tuner and tape deck or recorder. Frequency response: 76 20,000 Hz + 3dB, Output: 2 watts r.m.s. per channel into 8 ohms Inputs: phono 80mV; tuner/aux 80mV. Size 8" wide, 21" high

Q 25-WATT 3-WAY CRYSLER LIVING AUDIO SPEAKER CE-Sh

This high quality speak er has its own built-in 3-way sound response switch. giving you the ideal frequency response for -fi, natural or mood music listening, Its

£39.95 each. beautiful, heavy, oiled walnut cabingt incorporates two separate speaker units an 8" woofer, mid-range with 2" concentric tweeter. Power handling 25 watts r.m.s into 8 ohms. Overall frequency res

ponse! 35-20.000 Hz. Cabinet sizm: 10½" × 7½" × 8¾". Exactly right for matching the most modern decor



PALACE AM/FM/MPX STEREO TUNER AMPLIFIER SSA-16

This is one of the lowest priced staro tuner amplifiers on the market. It covers the full range of both AM and FM broadcast frequencies. And when you're switched to FM, an indicator lights up when a stereo signal is received – that's the time to switch to 'Stereo'! The SSA-16 has all the facilities you'd expect to find on tuners costing twice as much — separate vol-ume, bass, treble, balance and tuning controls. Selector switch one, oas, reure, dalance and cump controls, selector switch for trape, phono, AM, FM, stereo. Jack socket on front panel for stereo headphones. Frequency range: FM 88-108 MHz. AM 535-1605 kHz. Frequency response: 50-10,000 Hz \pm 3d8. Power output: 4 watts total music power into two 8 ohm speakers, Size: 16" wide, 4½" high, 8" deep.

ROC 7-WATT STERED AMPLIFIER CHASSIS SK-317 This exclusive R D C Stereo Chassis is self-contained, and it costs £2-25 less than the normal re tail value! The SK-317 is a really compact unit measuring only 5½" wide, 12"

high and 64" deep. It contains its own majns power supply, and has a ganged volume control and separate treble controls for each channel. Specification: frequency response 40-17 000 Hz + 3dB; output 3-5 watts music power per channel into 8 ohms; input, phono, 600mV; signal-to-noise ratio better than 45dB,



OLSON AM-372 16 WATT STEREO AMPLIFIER 🙉 Here's a really good amplifier at a really down-to-earth price – nearly £7 less than the normal retail value! Just look at what the AM-372 will do for you – reproduce signals from ceramic or crystal cartridges, AM and FM tuners, and tape recorders.

And it gives you outputs for two sets of speakers, headphones and tape recorders. Frequency response is 30 to 20,000 Hz \pm 3dB, Output 8 watts r.m.s. per channel music power into 8 ohm speakers. Phono input 200mV. Tuner input 200mV. Size $12\frac{\pi}{4}$ wide, $3\frac{\pi}{4}$ high, $7\frac{\pi}{4}$ deep.

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E18 E24 E12

E24 E12

4·7-470K 4·7-10M 4·7-10M 4·7-10M

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1Ω-10K E18 1Ω-10K E12

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628

THE BROADCAST BANDS

Malcolm Connah

Times in GMT Frequencies in kHz

NEWS FOR DX LISTENERS

T IS a generally accepted axiom that good shortwave reception is 50 per cent dependent on the aerial, the other 50 per cent accounts for the receiver and skill of the operator etc. I know many people who have bought expensive receivers and been disappointed with the results simply because they did not have an effective aerial system.

The simplest form of aerial is the long wire, and the longer the wire is the better the reception will be. The strength of the signal is measured in microvolts per metre; the longer the aerial is, the stronger the signal will be because the voltage induced in the aerial will be higher. All receivers have a certain sensitivity which is a measure of the voltage required at the aerial input to give a useful audio output.

A long-wire aerial of 50 feet will usually give adequate reception whilst one of 100 feet is very good. Other factors affect the performance of such an aerial, consequently the higher above ground it is the better it will perform. Care should also be taken to keep the aerial as far away as possible from buildings, trees and any other obstructions which may reduce the signal induced. It is also essential to keep the aerial as remote as possible from any sources of man-made interference.

A better type of aerial is the dipole. This type of aerial is restricted in use because it can only be tuned to a single band. The advantage of a dipole. however, is that being tuned its performance on the band of interest is superior to that of a long-wire.

The dipole aerial consists of a single length of wire which is cut in half, the two halves are connected via a coaxial cable to the two terminals at the rear of the receiver. The overall lengths for the various Broadcast Bands are as follows:

Frequency Band	Length of aerial
25 MHz.	5.60 metres
21 MHz.	6.65 metres
17 MHz.	8.00 metres
15 MHz.	9.30 metres
11 MHz.	12·10 metres
9 MHz.	14.80 metres
7 MHz.	20.00 metres
6 MHz.	23.40 metres

For the benefit of those who have not been converted to the metric system one metre is approximately equal to 3.28 feet (e.g. 20.00 metres = 65.6

An interesting feature of the dipole is that it . is a directional aerial. If the wire is aligned North-South the best reception will be from due East and

It is possible to use a dipole on all bands but without the same sensitivity as the band to which it

is tuned. For example a 7MHz, dipole will be a sensitive, directional aerial on the band of interest. but on all other bands it will perform as a nondirectional, 66 foot long-wire.

Most listeners have one band in which they are particularly interested, the ideal solution in this case would be to erect a dipole for this band and a longwire for general listening on the other bands.

The first report this month comes from Michael Berry in Dewsbury, Yorkshire. Michael is still using his Eddystone EB35 receiver but has replaced his vertical aerial with a 20 foot long-wire in the loft. The stations logged included:

3960 Baghdad, Iraq in Arabic at 2310.

4800 R. Lara, Venezuela in Spanish at 0240.

4820 Voz Evangelica, Honduras, English at 0340.

4880 R. Universo, Venezuela, Spanish at 0330.

4990 R. Barquisimeto, Venezuela at 0230.

4995 R. Brazil Central in Portuguese at 0110.

5075 R. Sutatenza, Columbia at 0100.

Richard Keen of Hayes in Middlesex has just erected a long-wire of 120 feet which enabled him to

4790 R. Ondas Portenas, Venezuela at 2055.

9510 R. Barquisimeto, Venezuela at 2015.

9595 R. Cultura de Bahia, Brazil at 2025.

9965 R. Lisbon, Portugal in English at 2050.

11875 R. Nacional de Nicaragua, Spanish at 2010.

12550 R. Nacional, Tanganyika at 2100.

John Adams of Sheffield returns to the column after a long absence. John has an Eddystone EC10 receiver, a Codar PR30X preselector and a 72 foot long-wire. His log included:

9570 Radio Australia with music at 0710.

9625 Kol Israel, English news at 2035.

11620 AIR, Delhi with news in English at 2008.

11650 R. Bangladesh in English at 1715.

13220 R. Euzkadi in Basque at 2135.

15018 Voice of Vietnam, Hanoi at 2005.

21495 R. Portugal with DX News at 0800.

Nigel Knowlman of Cullompton in Devon has sent in another good log which includes:

9570 Radio Australia in English at 0830.

9912 All India Radio in English at 2000.

15105 Radio Japan, English news at 0945.

15165 R. Denmark in English and Danish at 1200.

17655 Radio Cairo, 'Arabic by Radio' at 1800.

21535 RSA, South Africa, English news at 1400.

21545 Radio Ghana in English at 1440.

Reports should arrive by the 15th of the month and be addressed to me at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.

UN THE



from GW3ZSV. He tells enthusiastically of an 80 metre contest to be run by the Sully & District Short Wave Club on October 22. First section from 0900-1100 and the second from 1700 to 1900. Great news—the contest is open to both licenced Amateurs and s.w.ls so there's no excuse for not having a go. Rules are just about the simplest and easiest to follow that I've ever seen. Why not send GW3ZSV a s.a.e. and get all the gen? Address is 3 Thorley Close, Cyncoed, Cardiff, South Wales. For local s.w.ls with no club at present, the Sully gang meet every Tuesday at 1930 at the Sully Bowls Social Club. Sully is about 6 miles from Cardiff on the coast road to Barry. They cater specially for the s.w.l., too.

Elated squeaks from **Bill Waldron** (Cwmbran) whose 144MHz score since April 1 this year has now topped 40 counties and six countries. Bill mentions the infuriating situation during a recent 2 metre opening when London and Midland stations could be heard calling LA, SM, ON, OZ, etc., but only the faintest of whispers and bits of EU callsigns detectable down in Wales. He also remarks on hearing many QSO's of, say, Dudley talking to Wolverhampton and other local natters. He again asks why these locals don't turn their beams round towards GW. Almost every night, GW3ZTH, GW3LEW and GW3TMM are active on the s.s.b. channel. How about it you northern types—or are those beams rusted in position from the habit of local gossips?

On 80 metres, Bill complains of high static levels but still managed to log some s.s.b. from; EL0H/MM, FP0BG, OA4AS, PY2RT, PY3CGP, VE3GCS, VO1GQ/M, VP7ND, VP9BK, WA2FCA, W4NJF, W8BT, 5Z4MO, 8P6SJ (Scout Jamboree station).

Pretty stamp and long epistle from Victor Springer (St. Lucy, Barbados) relates hair-raising tales of s.s.b. received on a receiver using battery valves and a 90V h.t. line. Modifications include items like half of a 12AU7 as an outboard tuneable b.f.o. whose signal is injected by winding a few turns of wire around the first DF96. (Transistor types will doubtless find all this way above their heads!). (No, a 12AU7 hasn't got a collector). Heard on 14MHz with the aid of 150ft of aluminium wire in an unspecified shape; CE3AJE, DJ1HP, DJ2WF, DJ8EG, F3OX, G3KTJ, G3XJN, G400, G6GA, GB3MKT, HB9KM, HC2JX, HK5ACP, I5MLI, JA1BRK, JA4YAR, OA4AEH, PY1AGO, VXVK2SG, VK3JW, VK9JW, XE1LAG, YK1AA, ZD3A, ZK2DX, 4Z4JX, 5V2AT, 5V4GE, 5V8AR, 7X7Y.

Chance to get yourself a nice QSL from the land of GC. **Ken**, **GC3GPL** (ex G3GPL) is now active on c.w. between 14·000MHz and 14·050MHz most evenings between 1900 and 2230 plus Sunday mornings. He has put up a special antenna which, he hopes, will give reliable short skip signals into the UK. He will QSL any s.w.l. who sends him a useful report on

THE AMATEUR BANDS David Gibson, G3JDG

Frequencies in kHz • Times in GMT

his signals. Address for those useful (note that word, not just a "Heard you 579 pse QSL) reports is Haut du Pre, Fern Valley, St. Helier, Jersey, Channel Isles.

Patrick Funnell (Nottingham) has an 8-element Yagi on two metres. This feeds a dual-gate m.o.s.f.e.t. converter which in turn feeds a PW Clubman receiver used as a tuneable i.f. covering from 2-4MHz. He comments that the Clubman is a very good receiver and includes a 2 metre log which mentions stations like; OZ1EP OZ4HW, PA0KOK, PA0LSG (using n.b.f.m.), PA0PMQ, PA0TLX (n.b.f.m.) plus numerous G stations. It is interesting to note the number of stations using narrow band frequency modulation (n.b.f.m.).

Log from the quill of Alan Smith (Lancs.) using a JR310, 60ft. long wire plus a E-ZEE Match a.t.u. He complains of "... a combination of night starvation and v.f.o. tuners cramp". (Wait till you get knob nippers gout, that's when we really separate the men from the boys). Alan's log for 14MHz reads; CE3AQW, CE3OE, CO8GL, CP1JV, EA8EN, EP2TW, FC9VN, FM7WN, FY7AE, HC4BS, HK6CBS, HP1XIS, MP4TDM, OA4LV, OA4WJ, OD5AU, OX3HV, OX3JW, PJ2CC, PJ9AD, PZ5CW, TI2AS, VE1TB, VE7BNJ, VE8RCS, VK2QM, VK3DO, VK4CZ, VK5ID, VK7AK, VK0RC, VP2VAM, VK6FD. VU2BX, VU2ALU, XE1DO. XE1EH. VS6DO, XE3RE, ZL1VY, ZL2CB, ZL3JD, ZL4BC, ZM2ASJ, ZM4CR, 9K2AR, 8P6BC.

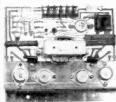
DX on ten metres is a GW, claims Martin "Squiggle" whose second name I can't decipher. (It's when your own g.d.o. is d.x. you start worrying!). Treasured possessions include a 640 receiver, a 20 metre dipole and a "flimsy bit of co-ax". Mr. Squiggle comments on the d.x. stations claimed by users of simple t.r.f. receivers, headphones made of draught excluders and coat-hanger aerials. He wonders if they use an SB303 receiver as a pre-amp and just forget to mention it. (Play fair lads, if you use a 30 transistor pre-amp, you gotta say).

A 9R59DE, 7MHz dipole and **B. Mainwaring** (Walsall) raised this lot on 21MHz; CE3OE, CP1HG, CR6CA, EA3VC, EA6BJ, EP2TC, JA1MTP, JA2MJ, JH1TMR, JH3ONP, OD5HF, OE6PN, PA0WF, PY1MT, SM4APD, 4Z4MO, 7X7Y. Interesting to note the European stations here (many haven't been included). Many s.w.ls commented on the short skip conditions which prevailed this past month. Effects like G stations on 20 metres being 5 and 9 were common. Wonder what things will be like next month, and how do you think the bands will fare this coming Winter? How about a line airing your views?

Logs, in alphabetical order please to arrive by the 15th of the month to 12 Cross Way, Harpenden, Herts.

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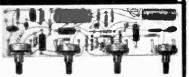
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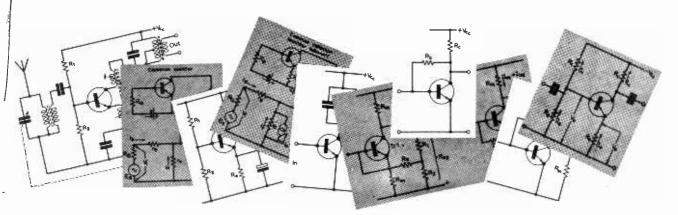
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PART 12

Virtual Earth in practice

Theory is all very well; an interesting intellectual exercise. Sooner or later, we must put our theories to the test, our principles into practice, and it is then—as any constructor will tell you—that the moment of truth comes. When you switch on, hold your breath, tentatively measure, go back meticulously over every step you have made and find the snags. That feeling of utter elation when a self-designed, self-made circuit bursts into life has no parallel in the world of module-massing. Humble though our circuitry may be, we are tempted to say: 'A poor thing, but mine own . . .'

So this month, following a discussion of virtual earth principles that may have left some of you

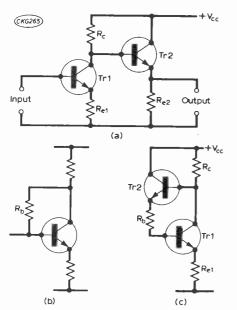


Fig. 61 (a): Redrawn circuit of Fig. 53 with some components omitted for simplicity. (b) base bias can be derived this way, included in (c) we have redrawn version of (a).

wondering what all the fuss is about, we put our theories into practice. Again, we shall proceed step by step, explaining *why* each component has to be its particular value, and what it does. Only this way is it possible to adapt ideas to your particular situation, i.e., to make your own design.

Part 10 in September PW, left us with a directly coupled pair of transistors, the first device acting as a collector-follower and with a potential divider for the derivation of its base bias. The second transistor operated in the common collector mode—that is, it was an emitter follower—and its base bias was derived from the collector of the first transistor, to which it is, (see Fig. 53, page 429, September PW) directly connected.

In the subsequent issue, under the guise of giving us all a breather, we talked about virtual earth circuitry, but you will have noted, in Fig. 60, that we again used a pair of transistors, intimately connected, directly coupled, so that their combined operation can be regarded, with some provisos, as that of a single unit.

Now let's take a look at a redrawn circuit of our earlier arrangement. Fig. 61 (a) is the redrawn circuit of our previous direct-coupled pair, this time with some components removed for simplicity. These are the base bias resistors, the input and the output coupling capacitors.

We still need base bias current, and this time we get it from the emitter of the second transistor. The important component is R_b. Take a look at another way of presenting such a circuit, Fig. 61 (b), which is really a redrawn Fig. 51 (*PW* Aug. 1972): now compare this with (c), which is, of course, a redrawn version of (a). The essential modification is that the base-emitter junction of the second transistor is included now in series with R_b. The similarity of the circuits is significant: similar spreads can be expected.

Gain

With this circuit, just as with the circuit of Part 10. gain is determined by the ratio of collector and emitter resistors of Tr1, which is to say, Rc1/Re1.

But this gain is before any feedback is applied. It is called the OPEN LOOP GAIN, and is denoted by the term G_0 . As we shall see later, though it can be calculated fairly enough, it is not quite so easy to verify by practical measurement.

When we put a bit of negative feedback in circuit, conditions alter. We are now taking a portion of the output and applying it to the input of our amplifier in such a way as to reduce the gain in proportion to the growth of the signal. If we apply the same feedback another way, of course, we can cause the amplifier to increase its gain in accordance with the signal, and thus go haywire. We are concerned here with negative feedback, reducing the gain selectively. So we get CLOSED LOOP GAIN.

The determination now is from the value of that feedback and the source input. In plain symbols the ratio of feedback resistor to source resistance, or

$\frac{R_b}{R_1 + \text{source resistance}}$

That magical term (if you read the right brochures!) the 'Virtual Earth' occupied us last month and we shall now return to it, as in Fig. 62.

Virtual earth point

Rource

Fig. 62: Virtual earth amplifier with feedback emitter Tr2 to base of Tr1.

Here, the virtual earth point is indicated and we must remember that the effective resistance at this point is determined by the ratio of that feedback resistor R_b and the open loop gain of the stage. This is R_b/G_o . Take a close look at this circuit, study it so that it can be recognised immediately, for this virtual earth circuitry will not show R_v , even dotted, as we have, and it is very easy to miss the implication. So let's now start to calculate the component values, for a practical circuit that can be part of a mixer, a microphone pre-amplifier, a variable gain pre-amplifier for an oscilloscope, and several other things.

Calculations

In Fig. 63(a) we have the same circuit, again shorn of unnecessary frills, and now we have a 'rail voltage', or h.t. positive, of Vcc, whichever term your age group has committed you to using, of 20 volts. The point is that we want a reasonable voltage 'swing' and if a collector is going to go up and down like a yo-yo, there will be a need for a high enough Vcc to accommodate this change in collector voltage. Thus far, we have been managing with 9-volt

batteries, convenient, easily obtainable, a constructor's aid. Now we must face the need for adding several in series or building a power supply operated from the a.c. mains. We shall look at the latter in a moment. For now, remember that the Vcc is 20 volts and for an emitter current of 2mA in the Tr2 circuitry, and a voltage at the emitter of about half the supply, we get $Re2=10/0\cdot002=5,000$ ohms.

Remembering that the base-emitter voltage will normally be 0.6V for the BC109, we say the base voltage will be 10.6V, and this is also the voltage at the Tr1 collector. We have already noted that for good stability it is desirable that the voltage drop across Rb is small. Under these proposed conditions, Tr1 would be operating with a collector to emitter voltage of a mere 600mV.

Now it is perfectly possible to use a transistor under these conditions, but the signal swing will be somewhat limited, so we look for an alternative way of doing things. One method is to split the feedback point, as in Fig. 63(b). Rel is maintained at the same ohmic value, but the feedback is tapped off

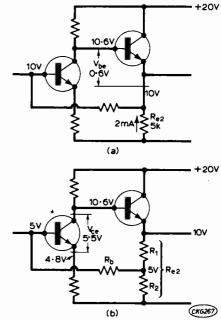


Fig. 63 (a): V.E. circuit redrawn. (b) tapping off of feedback point.

from the junction of the divided Re2, so that 5 volts is available at the base of Tr1 as bias. The voltages at Tr2 base and collector have not altered, so we have the condition now where Tr1 operates at a collector-to-emitter voltage of something like 5.5 volts.

Again allowing a 0.6 volts difference between base and emitter, we can now calculate Tr1 emitter resistor. The current flowing through Re1 can be large, using a small value of resistor, but this will not help us guard against noise. On the other hand, if we aim at a very low current, and hence a large resistor, our open-loop gain goes for a chop! Take an example: if the emitter current of the first transistor were to be 100 microamps, then Re1 would be $48k\Omega$ and Rc1 would have to be something like

94k Ω . From our earlier formula, $G_o = \frac{R_c}{R_e} = \frac{94k\Omega}{48k\Omega}$,

i.e., an open loop gain of something like two!

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MAGRETI A magnificently activated switch.
Vacuum sealed in a glass envelope. Silver
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Amps. 2 Amps. 2	
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1A MT 79 AT 1 (1.55 PPp 4A MT 21 AT 8.10 40p	
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1A MT 79 AT 1.55 20p 4A MT 21 AT 3.10 40p 2A MT 3 AT 3.25 30p 5A MT 51 AT 4.31 42p 50 volts, All tapped at 0-19-25-33-40-50V.	
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1A MT 79 AT \$1.85 369 4A MT 21 AT 3.10 409 2A MT 3 AT \$2.30 5A MT 51 AT 4.31 429 50 mA MT 102 AT \$1.45 76 3A MT 105 AT 3.31 41 1A MT 103 AT \$1.45 76 3A MT 105 AT 3.31 41 1A MT 104 AT \$3.50 3A MT 105 AT 3.50 44 2A MT 104 AT \$3.50 329 4A MT 107 AT 7.47 506 60 Volts. All	
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PARTRIDGE **ELECTRONICS**

23-25 HART ROAD, BENFLEET ESSEX SS7 3PB

We can, true, bypass the first emitter resistor with a whacking great capacitor, but let's not complicate matters at this stage. I hope the point is gradually being made that in this type of deceptively simple transistor circuit design one has to consider both a.c. and d.c. conditions; one can affect the other. What may be fine from the bias point of view may not be the best for signal conditions. We must press on ...

In practice

The practical answer to the problem is to use the potential divider again, but this time as shown in Fig. 64. This circuit is more like the real thing, as our airy-fairy calculated values of components now give way to preferred values, such as $4.7 \mathrm{k}\Omega$ for Re2.

Calculating the rest of the values requires recourse to the graphs for the particular transistor. If you have been following this series—and our mailbag shows, bless you, that quite a few have—you will be able to turn up the June issue and refer to Fig. 42, which was the $h_{\rm FE}$ against $I_{\rm c}$ curve for a BC109. We ignore the illustrative co-ordinates this time, and find that for an $I_{\rm c}$ of 2mA, the $h_{\rm FE}$ is 440,

so our base current of Tr2 is $\frac{2\text{mA}}{440}$ or around 4.5 microamps.

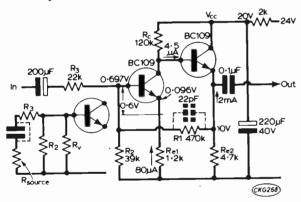


Fig. 64: Working circuit with component values.

As this current is flowing through the collector load of the first transistor, and is very small in comparison with the collector current of Tr1, we can safely disregard it in our next calculation. The emitter voltage of Tr2, we have said, is 10V, and the base will be 0.6V greater, so this 10.6 volts will be the collector voltage of Tr1. Therefore, the voltage across Rc1 will be the difference between the supply of 20 volts and this 10.6V, i.e., 9.4V.

From what we have already said about the choice between noise figures and signal swing at Tr1, we realise that there are limitations, upper and lower, for the emitter current, and for the balance between the feedback resistor and R2, now part of the bias circuit. Practically, our limits are from a minimum of 3 microamps and an upper of 200 microamps, which gives us R1+R2 of between $47 \mathrm{k}\Omega$ and $3 \cdot 4 \mathrm{M}\Omega$.

Plenty of choice? Maybe, but for a compromise value of emitter current that gives us preferred values of resistors and practical conditions, we have plumped for a figure of 80 microamps, as shown in

Fig. 64. So R_c becomes $\frac{9.4}{80 \times 10^{-6}}$ or approx. 120kΩ.

An open loop gain G_o of 100 is a healthy figure and as this is $\frac{R_c}{R_e}$ so R_e becomes $\frac{120k\Omega}{100}$ or $1\cdot 2k\Omega$.

If we again ignore the very small base current of the transistor, we can take it that the emitter current of Tr1 will be very near 80 microamps, and the voltage drop across Re1 will be 0.096V, which gives us a base voltage of nearly 0.7V. Referring again to the graph, we find for an Ic of 80 microamps we get an $h_{\rm FE}$ of around 280. So the base current will

be
$$\frac{I_c 1}{h_{FE} 1} = \frac{80 \text{ microamps}}{280}$$
 or 0.2857 microamps.

If the base current is to be provided by the potential divider R1 and R2, provided the current through these resistors is much greater than this, then Ib1 can be left out of subsequent calculations. But too low a combination of these resistors will affect the bias conditions of Tr2 (since we shall be shunting Re2 with the R1+R2 network). If we aim at a sum of the two at least 10 times that of Re2, we shall be safe. This summed resistance would be $47k\Omega$, and the current through them would be 200 microamps approximately—nearly a thousand times Ib1.

As to those limits: the minimum current we can allow through R1+R2 will be ten times Ib1, around 3 microamps, and the sum resistance would be $3\cdot 4M\Omega$. We have a very wide choice of values, but experience tells us that we cannot dip indiscriminately into the spares box.

TO BE CONTINUED

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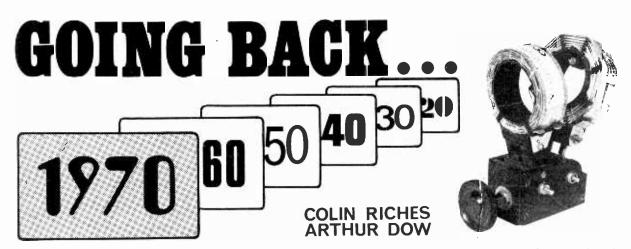
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FIRST, we would like to apologise if anyone has written to us and not received a reply but from correspondence we are now receiving, it appears as though a few letters have gone astray. If, therefore, you have not heard from us and would like to repeat your letters, we will be pleased to reply as soon as possible.

We have received quite a large number of "points from the post" that we feel merit publishing almost as they have been written, so, without any more

ado, let's begin . . .

R. L. Graper, 16 St. John's Avenue, Chelmsford,

Essex tells us . . .

"In June 1922 I bought the first issue of Popular Wireless . . . No. 1, Vol. 1, and later, also the first issue of Television, No. 1, Vol. 1 in March 1928. I kept them and still have them (a little soiled) in the hope that they might later become valuable as "collectors items". As the address of Popular Wireless is the same as that of your present premises, I am wondering whether you took over the above journal.

There is much of interest in this first issue, and I

quote you a few items as follows.

"Have you a wireless? First take out a licence. In a week or so this may be obtained direct from any post office. At present you must apply to the Secretary of the G.P.O. Cost of licence . . . 10/-.

Article. "Mr. Selfridge expresses his views. The

business man and radio, by G. Selfridge."

I do not think the radiophone will ever supercede the telephone. Many wild predictions have been made on this point, and it has even been said that before long the man in the street will have his pocket transmitter with which to call up his friends or business colleagues. That prediction is very far fetched indeed. The pocket radiophone will never replace the telephone, because it will always be limited by the number of messages that can travel through the ether at the same time. At the moment, American amateurs are realising the bad effects of unlimited radio transmission . . . The ether is crammed with their messages, which consequently get so mixed up that it is impossible to hear a coherent sentence when listening in".

Short NEWS FLASH . . . Mr. Marconi's voyage . . . Mr. Marconi proposes to carry out experiments on the Atlantic with direction finders on the Short-wave, and the Long-wave transmission. Besides his other experiments, Mr. Marconi will carry out tests for the Meteorological Office. The "Electra", Mr. Mar-

coni's steam yacht of 700 tons, is the finest equipped floating wireless laboratory in the world."

Interesting prices . . . Instrument wires. Aerial Wire 7/23, which was of course seven strands of 23 swg. copper wire. Price: 6/6 per 100 feet. Double silk covered wire 12 swg: 4/- per lb, Single valve Detector panel: 32/-, single valve h.f. panel: 35/-. These two panels together gave, it was claimed by the advertiser, a complete two valve receiving set. This was capable of receiving all music, speech etc. from all British stations.

In my copy of Television No. 1, March 1928, there are many interesting items, such as the early experiments by Baird, with a short paragraph giving details of his first demonstration before the Royal

Institution.

There are pages of advertisements for (guess what) Radio sets and parts for the radio constructor. As no Television receiver is shown or advertised, I presume this came a few months later. The first experimental transmissions on the 30 line system could be received on home made apparatus, and there are articles (most detailed) on Disc Receivers, Light sensitive cells, a cartoon drawn specially by Heath Robinson, and several articles on the future miracles of this new science . . . which indeed have in some cases become fact today.

In this year there were quite a number of Valve Portables, then priced in the £30 region, but one would rather call them "transportables" I think.

Vintage Equipment

Mr. G. H. Wallace was recently presented with a pair of old headphones and he would like to know what impedance they are and what vintage they are likely to be.



continued on page 641

BSR LATEST SUPERSLIM STEREO AND MONO

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Auto or Manual. A high quality unit backed by BSE reliability with 12 months' guarantee. AC 200/250v.
Size 13† × 11†in.
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with STEREO and MONO XTAL MONO-COMPATIBLE



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£10 Post Free 4 speeds, Plays all sizes of records. 4 pole heavy duty motor. 9in. steel



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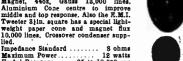
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2000mF, 6V, 25p; 25V, 42p; 25V, 75p; 55V, 85p; 50V, 95p,
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EF183	12 p	PCC89	12 i p	U191	17 tp
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EY86	17 t p	PCL82	17 p	30PL1	22 i p
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GOING BACK —continued from page 638

They are about 2^{1} ₈in. diameter and constructed from aluminium. We have persuaded the great "PAX" of "Practically Wireless" fame to switch off his vintage Hawaiian 78's for a moment and do a sketch to show how Mr. Wallace's headphones look.

If you have any information on these, please send it to Mr. Wallace at 3 Woodside Avenue, Kilmarnock, Ayrshire, Scotland.

Miss C. Crabb writing from Bournemouth, Hampshire, tells us that she recently unearthed the receiver and speaker shown in the photographs.

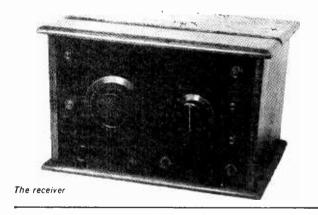
The receiver was licensed under the Marconi patent and has reactance tuning. The grid bias battery is fitted inside the cabinet.



Miss C. Crabb's horn loudspeaker

Miss Crabb asks if this equipment is of any value as a curio or as a collectors' item.

In answer, we can only stress once more that equipment of this vintage (and indeed any equipment pre-1930) is certainly worth hanging on to. Collectors would put a value on it and even the person wanting such an item as just a "curio to put on the shelving in the lounge" would possibly be prepared to offer a reasonable sum.



If any other readers have any information on this receiver and would care to write to us, we will be pleased to pass on the information to Miss Crabb.

Vintage Record Bazaar

We would like to thank readers who have written in letting us know of their old gramophone records and equipment. We hope to run an article in a future issue on early gramophones, and will find the information you have provided will help us greatly in our project.

Readers may also be interested to know that the 5th annual Vintage Record Bazaar will be held on Saturday, Oct. 14th, at St. Silas Hall, 74 Penton Street, Islington, London, N.1.

The first two of these record sales, five years ago were held in conjunction with vintage car rallies, there being 15 record stalls.

This year, there will be 50 stalls with over 30,000 vintage records on sale. Vintage post cards will also be in evidence if you are interested in collecting them. In addition, there is a strong possibility that there will be some vintage radio equipment there!

The Bazaar is open from 10 a.m. to 5 p.m. and admission is 25p.

If anyone would like any further information on the Vintage Record Bazaar, phone John Carter (Shurlock Row) 539, or if you are phoning from London, 0735-81-539.

ELECTRONOTES—continued from page 625

the physical properties of the medium at various depths.

These points are important since the acoustic signal pulsed out by the transmitter in an "active" system could be diffracted in a curved path and thus there would be certain completely blind areas.

The thermal profile of the water at various depths can be quite a complex thing. This profile will alter daily being dependant upon the day to day variation of the solar activity of the sun. Again, winds and rain cause a mixing action which is yet another variable.

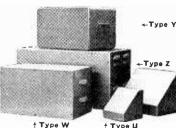
We have barely scratched the surface of this fascinating subject. However, it appears that we have a long time to wait before some boffin comes up with a useful sonar device for the accurate location of the soap in murky bath water. Or does some reader just happen to have a practical circuit handy?



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I.C. Stabilised Power Supplies—continued from page 625

Although component layout is unimportant two details must be observed. The voltage across C1 must he less than 36V before the i.c. is inserted into the circuit. If the voltage across C1 exceeds 36V this voltage can be reduced by altering the tapping of the transformer. Secondly it should be noted that the output power transistor will, under current limiting conditions, have to dissipate about 40W. An adequate heat sink for this component must be provided, the chassis of the supply being used for this

* components list

R1 2kΩ ½W 5% R2 560Ω ‡W 5% R4 1.5kΩ ½W 5% R3 0.5Ω 3W wirewound VR1 10kΩ skeleton preset potentiometer C1, C2 2500 µF 40 VW minimum. C3 0-01 µF

Tr1 2N3055 Tr2 2N3704 Tr3 2N3704 D1-4 3A silicon rectifiers or bridge D5 BZY88C9V1 (9-1V 400mW) D6 BZY88C6V2 (6·2V 400mW) ICI HA741C T1 230V/24V 1.5-2A

Red and black insulated terminals. Neon lamp and holder 230V. Fuses 2-5A and 1A with holders. Meter 50VDC. Mains input plug and socket. On-off switch. Case and chassis. Veroboard 41 x 4in.

The components list above applies only to the variable voltage supply. The veroboard used in the prototype has pins fitted which plug into a chassis mounted socket.

purpose. The transistor is however electrically isolated from the chassis by using a greased mica washer and bushes. Although no other difficulties should arise it is recommended that the semiconductors, and especially the i.c. should be guaranteed to be to the full specification. Providing all these precautions are taken the supply should prove to be completely reliable.

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Sunderland SR4-OBB.

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1970. Also P.W. April—Dec. 1968.—V. Simpson, 107 Derryhale Road, Portadown, Co. Armaph, NI.

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...wanted to correspond; Someone of my own age (17). Interested in Electronics and Modern Music.—P. Richards, 7 Holwood Drive, Whalley Range, Manchester M16 8WS...girl or boy of my own age (14) to correspond or tapespond with. have a Prinsound TR-5 1 /8 cassette recorder and a Ferguson 3 ½ and 1 ½ spool recorder with 4 tracks. I am interested in electronics and SWLing.—K. Goldsworthy, 165 Ellerdine Road, Hounslow, TN3 2PV.
...boy or girl of my own age (16). Anyone who is interested in TX circuits and short wave listening.—P. Jenkins, 30 Gainsborough Road, North Finchley, London, N.12 8AG

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.anyone who is interested in offshore radio and radio equipment to correspond. Aged 18-24.—P. Leach, The Post Office, Snodland, Kent.

.Elther sex of my own age (24). I have a four track Izlow AKAL stereo tape recorder and interested in modern music, radio electronic, and pen friendship.—M. Ekramul Haque, BBC eastern relay station, BFPO 65, Via London.

.boy or girl of my own age (13) interested in Radio and Electronics.—D. Thorpe, 161 Tomswood Hill, Hainault, Illord, Essex.

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.anyone of about my age, 17, interested in amateur radio and S.W.L.,—P. Jenkins, 30 Gainsborough Road, North Finchley, London, N.128AG.

.I would be delighted to correspond with any radio constructor from England, of my own age (25).—K. H. E. Perera, 352 Kerawalapitiya, Hendala, Wattala, Ceylon.

.girl of my own age (almost 13) interested in Electronics and pop music.—L. Cox, Ford Bank Cottage, Victoria Avenue, Didsbury, M20 8RA.

.I would like to correspond with radio constructors from England, of my own age (18) S.A.M.P. Ratnayaka, no 330 Rajamaha Vihara Road, Mirihana, Pita-Kotte, Ceylon.

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.Anyone interested in sending me Radio parts against anything of equal value from India.—S. Khaitan, 25/1 Rowland Road, Calcutta 20, W. Bengal, India.

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TAPESPONDENTS WANTED
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Peters. Liverpool Street, P.O. Box 6, Eket, S. East State of Nigeria.
..boy or girl 15-18. Cassette or open reel catered for. My interests include music, tape recording (1) and electronics. Please write stating interests, recording equipment etc.—M. Tillman, 86 Chestnut Drive, Pinner, Middlesex. HAS ILZ.

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... Ferrograph/3CFN/X tape recorder for sale, good running order. Any offers.—S, G. Parsons, 7 Randolph Terrace, Stirling, FK7.7AB.
... FOR SALE; Five old valves (from Portadyne 230 radiogram) all In good condition, probably working, All marked very clearly, 5, 6, 7, & 8 pin bases. Any offers.—M. Baldwin, 8 Oakcroft Road, London, SE13.7ED.
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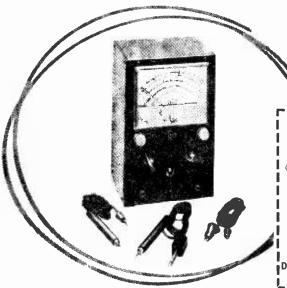
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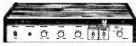
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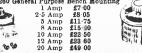


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0.20 0.20 0.18 0.10 2N3904 2N3905 2N3906 2N4036 2N4037 2N4058 2N3059 2N4060 2N4061 2N4209 0·10 0·11 0·10 0·09 0·10 0·80 2N2219 0·18 0·15 2N2219A 2N2220 BF185 C425 C426 C428 C444 D40N3 GET111 GET113 GET114 GET115 2N2221 BC268 BC209 BF194 0.12 BF195 BC209 BC211 BC212K BC212L BC214L BC236 BC237 BC238 BC239 BC251 0.42 0.47 0.17 0.15 0.25 0.84 0.49 BF196 BF197 BF198 BF199 0·10 0·12 0·14 0·16 0·11 0·12 0·15 0·41 2N2369 2N4062 2N4302 2N4303 2N4913 2N4915 2N4916 2N4917 2N4918 2N4919 BF200 0·35 0·14 0·19 0·22 BF224J BF225J BF257 BF238 BF244 BF245 BF246 BF247 0.08 0.08 0.08 2N2646 1.20 0.12 0.12 0.17 0.17 0.18 0.25 0.26 2N2647 0.85 2N2711 0.50 0.45 0.50 1.28 0.55 0.38 GET120 2N2711 2N2712 2N2713 2N2714 0.60 0.22 0.16 0.88 0.48 0.49 GET535 GET536 BC251 BC252 BC253 BC257 BC258 BC259 BC261 BC262 BC263 0.20 0·18 0·23 0·08 0·08 0·11 0·20 0·18 0·23 0.20 GET538 0.20 ⁷290 **GET873** 2N2904A 0.12 2N2905 2N4920 BF254 0·14 0·15 GET880 2N4920 2N4921 2N4922 2N4923 2N4926 2N4927 2N4928 2N2905A GET883 BF255 0.50 0.60 1.50 1.75 0.24 0.35 0.25 0.24 BF257 0.41 GET887 2N2906A 0.23 BC300 0·42 0·34 BF258 0.46 **GET890** 0.22 BF259 GET895 2N2907A 0.25 BC302 0·27 0·50 0.28 1.45 BF261 MA8001 0.16 2N2913 0.75 2N4929 2N2923 0.12 2N4930 BC303 MA8002 BC304 0.43 BF270 0.25 MA8003 0.47 Post & Packing 13p per order, Europe 25p. Commonwealth (Air) 65p (MIN.) Matching charge (audio transistors only) 15p. extra per pair. Prices subject to alteration without prior notice.

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SN7407	0.30	0.24	SN7440	0.20	0.18	SN 7480	0.70	0.65
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SN7412	0.42	0.86	SN7445	1.85	1 70	SN7485	2.50	2.80
SN7413	0.80	0.27	SN7446	1.00	0.90	SN7486	0.83	0.28
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SN7417	0.30	0.25	SN7448	1.04	0.92	SN7489	4.97	4-40
SN7420	0.20	0.18	SN7449		e to be	SN7490	0.75	0.70
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1 A	8p	9 p	10p	11 p	12p	15p	20s	
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1 amp an	d 3 amp a	re plast	ic encar	sulation.				

		DIODE	ES &	RECTI	FIERS		
IN34A IN914 IN916 IN4007 I844 IB113	10p 7p 7p 20p 7p 15p	AA119 AA129 AAZ13 AAZ15 AAZ17 BA100	7p 15p 12p 12p 10p 16p	BAX16 BAY18 BAY31 BAY38 BY100 BY103	12ip 17ip 7p 25p 15p 22p	FST3/4 OA5 OA10 OA9 OA47 OA70	2219 17p 20p 10p 8p
I8120 I8121 I8130 I8131 I8132 I8920 I8920 I8922 I8923 I8940	12p 14p 8p 10p 12p 7p 8p 12p 5p	BA102 BA110 BA114 BA115 BA141 BA142 BA144 BA145 BA164 BAX13	25p 25p 15p 7p 17p 17p 12p 17p 12p	BY122 BY124 BY126 BY127 BY164 BYX10 BYZ11 BYZ12 BYZ13	47 p 15p 15p 17p 57p 22p 35p 82p 80p 25p	OA73 OA79 OA81 OA82 OA90 OA91 OA95 OA200 OA202 TIV307	10p 79 89 10p 79 7p 7p 7p 10p 50p

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SINGLE GANGED, LOG or LIN
lk to 1M. 40p each.
TWIN GANGED, LOG or LIN
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MULLARD C280 M/FOIL

-		JNJ		
0.01, 0.	022, 0	.033,	0.047	Sp each
0.068, 0				4p each
0.15, 0.2	22, 0.3	3		5p each
0.47				97
0.68				119
lμF				149
1.5µF				21p
2.2/4F				25 B

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AC142K			8p			OC201		2N3055	49p	1N4002	4p
AD14		BC154	20p	ME0404	14p		25p				
AD150	44p	BC168	10p		10p	OC25	25p	2N3702	12p	1N4003	5p
AD161		BC169	11p	ME4102	12 _D	OC28	30p	2N3703	12p	IN4004	7p
AD161 M	P55p	BC182L	8p	ME6002.	14p	OC29	36p	2N3704	12p	OA90	бр
AF114	15p	BC183L	8p	ME6101	14p	OC35	25p	2N3705	12p	OA91	βр
AF115		BC184L	8p	ME6102	15p	OC36	36p	2N3706	10p	OA200	10p
		BC212L	8p	MP8111	32p	T1129A	48p	2N3707	10p		8p
AF116										1844	10p
AF117	15p	BC214L	8p	MP8511	84p	TIP30 A	55p	2N3708	9p		
				MP8513	45p	TIP31A	58p		10p	IN4149	4p
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ı	=8N7402	0.15	=8N7473	0.37
ı	=8N7403	0.15	=8N7474	0.37
	=8N7404	0.15	=8N7475	0.47
ı	=8N7405	0.15	=SN7476	0.48
ł	=8N7407	0.18	=8N7480	0.67
	BN7408	0.18	=8N7481	0.97
	=8N7409	0.18	=8N7482	0.97
	=8N7410	0.15	= 8N7483	1.10
	=8N7413	0.29	= 8N7486	0.82
	=8N7416	0.48	=8N7490	0.67
	=8N7417	0.43	=8N7491AN	0.87
	=8N7420	0.15	=8N7492	0.67
	=SN7430	0.15	=8N7493	0.67
	=8N7440	0.15	=8N7494	0.77
	=8N7441	0.67	=8N7495	0.77
	=8N7442	0.67	= SN7496	0.77
	=8N7443	1.95	=8N74100	1.75
	=8N7444	1.95	=8N74104	0.97
	=8N7445	1.95	=SN74105	0.97
	= 8N7446	0.97	=8N74107	0.40
	=8N7447	0.97	=8N74110	0.55
	=8N7448	0.97	=8N74111	1.25
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0.15

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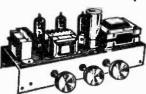


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6AQ5	21 6J7G	.24 12AV6	.28 35 A 3		CC 1-65		. 54	HVR2	. 58	PL36	. 46	UF80 .35	AC113	. 25	BCY12	. 50	OA79	. 09		. 58
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Take flash pictures at the exact instant of the bursting of a balloon. A champagne cork leaving the bottle. The split second a harmore strikes a light bulb. The mind bogdes at the possibilities of the "Phototron". The only timit is your imagination. Now that inexpensive flash gare on the market in quantity, it has made possible, with the help of electronics the production of a wide variety of exciting photography effects, once strictly limited to the professionals. The duration of an electronically produced shab is extremely brief and normally measured in Millisconds. Now, providing the camera shutter is left open in a dark or subdued light, it is the timing of the flash that determines what is imprinted on the film—not necessarily anything done by the camera. As electronic flash guns are fired by making a switched connection then it becomes obvious that one of the latest Silicon Controlled Relay's can be used. If we make this S.C.R. operate by sound (with variable sensitivity control for soft/houd sounds) some real "way out" photography effects can be captured forever on fline. Easily built in a coujle of hours or so, the "Phototron" is fully solid sate, uses self contained PPS battery. No soldering necessary using our special printed circuit terminal board (pat. Applied for). All parts including special pictorial step-bystep plans, transistors, microphone, S.C.K., potentiometers, switches, test lamp, case, not separately).

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tracks down burled metal objects—It signals exact location with loud audible sound (no phones used)—uses any transistor radio which fits inside—no connections needed. FINDS GOLD, SILVER, COINS, JEWELLERY, ARCHAEOLOGICAL PIECES ETC. Extremely sensities will signal presence of certain objects buried every signal presence of certain objects buried ground. No known pround. No known pround pround pround pround pround pround pround pround pround. No known pround pro

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Anyone from 9 year up can follow the step-by-step, easy as ABC fully illusstrated instructions.

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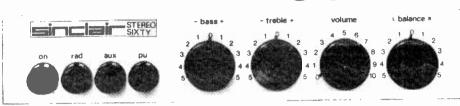
Actually whistles & warbles like a real live Canary! Amazing circuity fall-fully reproduces the Canary's magnificent song. Just switch on and lead it—whistling down scale for some seconds, suddenly breaking into a delighbol warble, then several seconds later shutting off for a second or two—only to start automatically again in a few seconds. People listen to the delightful joy to young and old alike. Standard self-contained battery lasts ages. Easily built in an hour or so with special pictorial step-by-step plans. No soldering necessary. All parts including printed circuit terminal board (pat. applied for) transformers, loudspeaker, transistors, nots, serews, wire, etc. + case (which its under cage—cage and toy bird not supplied) ONLY 24-50 + 25p. P. & P. (parts available separately).

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Sinclair Project 60

Stereo 60



Built and post free f9.98

pre-amplifier/control unit

The versatility of Project 60 high-fidelity modules is well demonstrated in this excellent unit. It provides the facilities essential to good stereo and will enhance the quality of any system it is used with, whether Project 60 or any any other top line power amplifiers. Compact, yet robustly constructed, the unit is easily panel mounted and will operate satisfactorily from 18 to 35 volts supply. Silicon epitaxial transistors are used throughout to achieve a very high signal to noise ratio with excellent separation between channels. Distortion at maximum output is barely 0.02% with magnetic p.u. input. Accurate equalisation is provided for all inputs, which are selected by push buttons. For maximum effectiveness, the Sinclair A.F.U. is recommended for use with the Stereo 60 pre-amp/control unit. A comprehensive manual supplied with Project 60 modules makes installing and connecting easy and ensures best possible results from your system.

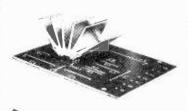
SPECIFICATIONS

Input sensitivities: Radio – up to 3mV Mag p.u. 3mV correct to RIAA curve – 1dB, 20 to 25,000 Hz Ceramic p.u. – up to 3mV; Aux – up to 3mV Output: 250mV Signal to noise ratio: better than 70dB. Channel matching: within 1dB

Tone controls: TREBLE + 12 to —12dB at 10KHz; BASS + 12 to —12dB at 100Hz Front panel: brushed aluminium with black knobs and controls

Size: 66 - 40 - 207mm.





aving introduced Integrated Circuits to hi-Instructors with the IC 10, the first time an Cad ever been made available for such purposes we have followed it with an even more efficient ersion, the Super IC 12, a most exciting advance ver our or gina unit. This needs very few ex ternal resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up, F.M. radio or small P.A. set up, etc. The free 40 page manual supplied, details many other applications which this remarkable IC make possible it is the equivalent of a 22 tran

sistor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC.12 is compatible with Project 60 modules which would be used with the Z 50 and Z.30 ampl fiers. Complete with free manual and printed circuit board

SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak) 6–8Ω Frequency Response: 5Hz to 100KHz-1dB Total Harmonic Distortion Less than 1%. (Typical 0-1%) at all output powers and frequencies in the audio band (28V) Load Impedance: 3 to 15 ohms. Input Impedance: 250 Kohns none as 1 states of 10 to 10 t pedance: 250 Kohms nom nal Power Gain | 90dB (1,000,000,000 times) after feedback Supply Voltage: 6 to 28V. Quiescent current: 8mA at 28V. Size: 22×45×28mm in clud ng pins and heat sink

Manual available separately 15p post free

With FREE printed circuit board and 40 page manual

£2.98 Post free

Project 605





Project 605 is one pack containing one PZ5 two Z30's, one Stereo 60 and one Masterlink. This new module contains all the input sockets. and output components needed together with all necessary leads cut to length and fitted with neat little clips to plug straight on to the modules. Thus all soldering and hunting for the odd part is eliminated. You will be able to add further Project 60 modules as they become available adapted to the Project 605 method of connecting

Complete Project 605 pack with comprehensive manual, post free

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Z.30 & Z.50 power amplifiers

Built, tested and guaranteed with circuits and instructions manual. **Z.30 £4.48 Z.50 £5.48**

The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to provide unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at 15w (8 Ω) and all lower outputs, Whether you use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and are intended for use principally with other units in the Project 60 range. Their performance and design are such, however, that Z.50s and Z.30 may be used in a far wider range of applications.

SPECIFICATIONS (Z.50 units are interchangeable with Z.30s in all applications).

Z.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts.

Z.50 40 watts R.M.S. into 3 ohms using 40 volts 30 watts R.M.S. into 8 ohms using 30 volts.

Z.50 40 watts R.M.S. into 3 ohms using 40 volts 30 watts R.M.S. into 8 ohms using 50 volts.

Frequency response: 30 to 300.000Hz ± 1dB. Distortion: 0 02% into 8 ohms. Signal to noise ratio: better than 70dB unweighted. Input sensitivity: 250mV into 100 Kohms (for 15w into 8Ω). For speakers from 3 to 15 ohms impedance. Size: 14 x 80 x 57mm





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Project 60 Stereo F.M. Tuner

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 plus 20 m.l.C. Tuning range: 87.5 to 108MHz. Sensitivity: ΛμV for lock-in over full deviation. Squelch level: Typically 20μV. Signal to noise ratio: > 65dB. Audio frequency response: 10Hz = 15KHz (±1dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: 2μV. Cross talk: 40dB. Output voltage: 2 x 150mV R M S. maximum Dperating voltage: 25–30VDC. Indicators: Stereo on ; tuning. Size: 93 x 40 x 207mm.



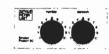
A.F.U. High & Low Pass Filter Unit

For use between Stereo 60 unit and two Z.30s or Z.50s. The unit is very easily mounted and is unique in that the cut-off frequencies are continuously variable. As attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. There are two filter sections - rumble (high pass) and scratch (low pass). H.F. cut-off (-3dB) variable from 28KHz to 5KHz. L.F. cut-off (—3dB) variable from 25Hz to 100Hz. Distortion at 1KHz (35V. supply) 0.02% at rated output. Operating voltage from 15 to 35V. Current 3mA. Size: 66 x 40 x 90mm



£5.98

£25



Power Supply Units

Designed specifically for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 or PZ.8 where a stabilised supply is essential

Typical Project 60 applications

PZ.5 30 volts unstabilised	£4.98
PZ.6 35 volts stabilised	£7.98
PZ.8 45 volts stabilised	
(less mains transformer)	£7.98
PZ.8 mains transformer	£5.98





System	The Units to use	together with	Units cost
Simple battery record player	Z.30	Crystal P.U., 12V battery volume control, etc.	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U volume control, etc.	£9.45
12W. RMS continuous sine wave stereo amp. for average needs	2 x Z.30s, Stereo 60; PZ.5	Crystal, ceramic or mag PU, FM Tuner, etc.	£23.90
25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers	2 x Z.30s, Stereo 60; PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc.	£26.90
80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms)	2 x Z.50s, Stereo 60; PZ.8, mains transformer	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43



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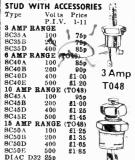
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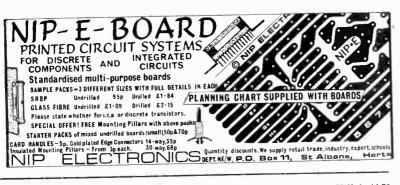
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OA2 0.40	6AR5 0.55	6F1 0.75	7Y4 0.75 9BW6 0.70	First Qua	lity F	ully Guar	anteed	EF97	0.65	GZ33 0.75		0.55		0.70
OA3 0.48	6AR6 0.65	6F5 0.75		LILAL Mad	iity i	uny waan	4111000	EF98	0.75	GZ34 0.60		0.50		0.80
OB2 0.40	6AR11 1.25	6F6G 0.45	10C2 0.60					EF183		HABC800.55	PL82	0.45	UABC80	- 1
OB3 0.70	6A85 0.50	6F11 0.50	10D1 0.65						0.85	HK90 0.50	PL83	0.45		0.40
OC3 0.40	6AB7G 0.85	6F13 0.50	10D2 0.55			A Day			1.25	KT66 2.35		0.40	UAF41 (0.70
OD3 0.40	6AT6 0.38	6F14 0.70	10F1 0.75		A # 1 # # # # # # # # # # # # # # # #	3.7		EK90	0.82	KT88 2:25		0.95		0.60
1B3GT 0.45	6AU6 0.80	6F15 0.65	10F9 0.65 10F18 0.60	14		BRA	ND	EL34	0.50	N78 1.60		0.75	UB41 (0.65
1L4 0.25	6AV6 0.40	6F18 0.50	10F18 0.60	140	aerv				0.50	PABC800.40		0.90	UBC41	0.55
1R4 0.50	6AW8A 0.65	6F23 0.90	10L1 0.60		1.00			EL37	1.70	PC86 0.60		1.10	UBC81	
1R5 0.45	6BA6 0.28	6F24 0.80	10LD11 0.70	A STATE OF THE PARTY OF THE PAR				EL41	0.75	PC88 0.60		1.00		0.40
184 0.80	6BE6 0.32	6F25 1.00	10P13 0.75				LVEC	EL81	0.55	PC92 0.05		0.95		0.40
185 0.80	6BF6 0.55	6F26 0.35	12AB5 0.70	ELEC'	r R O N	IC VA	LVES	EL83	0.50	PC97 0.50		0.60		0.70
1T4 0.30	6BH6 0.76	6F28 0.70	12AC6 0.60	ELLO	•		_			PC900 0.48		0.85		0.70
1U4 0.40	6BJ6 0.55	6GK6 0.60	12AD6 0.60				TGG04 0 00	EL84	0.25		PY33	0.68		0.45
1U5 0.75	6BK7A 0.75	6J4 0.60	12AE6 0.60		50EH5 0.65	DK96 0.60	ECC84 0.80	EL85	0.48	PCC84 0.40 PCC85 0.40	PY80	0.40		0.45
1V2 0.65	6BN5 0.43	6J5GT 0.40	12AL5 0.55		50L6GT 0.60	DL96 0.55	ECC85 0.40	EL86	0.40		PY81	0.30		0.70
1X2B 0.55	6BN6 0.60	6J6 0.30	12AQ5 0.50	30C15 1.00	85A2 0.55	DM70 0.60	ECC88 0.40	EL90	0.42	PCC88 0.55	PY82	0.85	UCH21	
2A3 0.50	6BQ5 0.25	6J7 0.45	12AT6 0.40	30C17 1.10	90AG 2.40	DM160 0.68	ECC89 0.50	EL95	0.35	PCC89 0.55		0.38		0.70
2CW4 0.75	6BR8 0.76	6K6GT 0.75	12AT7 0.40	30C18 0.90	90AV 2.50	DY51 0.55	ECC91 0.80	EL821	0.60	PCC189 0.80	PY83	0.40	UCH	
2D21 0.40	6B87 1.35	6K7 0.43	12AU6 0.45	30F5 1.00	90C1 0.75	DY86 0.85	ECC189 0.65	EL822		PCC805 0.95	PY88	1.00		0.60
	6BW6 0.90	6K8G 0.45	12AU7 0.88	30FL1 0.80	90CV 2.40	DY87 0.36	ECC807 1.00	ELL80		PCC806 0.95				0.85
	6BW7 0.90	6K25 0.75	12AV6 0.45	30FL12 1.10	807 0.50	DY802 0.87	ECF80 0.35	EM34	1.00	PCF80 0.80		0.47		
	6BX6 0.25	6L6GT 0.55	12AV7 0.70		813 4.00	E88CC 0.70	ECF82 0.35	EM71	0.80	PCF82 0.35		0.50		0.65
	6BZ6 0.45	6L7 0.45	12AX7 0.88		866A 0.85	E180F 1.00	ECF8041.65	EM80	0.45	PCF84 0.60	PZ30	0.88		0.00
384 0.40	6C4 0.35	6L18 0.50	12AY7 0.75	30L15 0.95	5642 0.70	E810F 2.90	ECH81 0.80	EM81	0.60	PCF86 0.60	QQVO2	1-6 0 0*	UF11	0.65
51:4GY 0.75 5U4G 0.40	6C5GT 0.55	6LD20 0.50	12B4A 0.65	30L17 0.95	5670 0.60	EABC800.88	ECH83 0.45	EM83	0.50	PCF87 1:10		2.25		0.65
	6CB6 0.40	6N7GT 0.55	12BA6 0.45	30P12 1.00	6080 1.75	EAF42 0.60	ECH84 0.45	EM84	0.35	PCF801 0.50	QQVO			0.65
5V4G 0.50 5Y3GT 0.45	6CD6GA	6P28 0.65	12BA7 0.50	30P19 0.95	6146 1.60	EBC33 0.60	ECL80 0.50	EM85	1.00	PCF802 0.50	00000	1.25		0.85
5Z3 0.75	1.30	6Q7 0.50	12BE6 0.50	30PL1 0.95	6146B 2.50	EBC41 0.65	ECL81 0.50	EM87	0.70	PCF805 0.90	QQV03			0.40
	6CG7 0.60	68A7 0.45	12BH7 0.50	30PL13 1.10	6360 1.25	EBC81 0.88	ECL82 0.35	EN91	0.40	PCF806 0.75	mmo1	5.25 3.40		0.40
5Z4G 0.45 6/30L2 0.90	6CH6 0.60	68G7 0.45	12BY7 0.65	30PL14 1.25	6939 2.25	EBF80 0.40	ECL83 0.70	EY51	0.40	PCF808 0.90	TT21			0.65
	6CL6 0.60	68K7 0.50	12K5 1.00	35.A3 0.60	7199 0.85	EBF83 0.40	ECL84 0.55	EY80	0.75	PCH2000.70	TT22	8.50		0.48
	6CU6 0.80	68L7GT0.45	12K7GT0.50	35A5 0.75	7360 2.20	EBF89 0.82	ECL85 0.55	EY81	0.40	PCL81 0.60	U18/20			
	6CW4 0.70	6BN7GT0.45	12Q7GT 0.45	35B5 0.65	7586 1.50	EBL31 1.50	ECL86 0.40.	EY83	0.55	PCL82 0.85	U25	0.85		0.80
6AG5 0.25	6CY5 0.50	68Q7 0.50	128 R7 0.50	35C5 0.60	7895 1.50	EC53 0.50	ECLL800	EY86	0.40	PCL83 0.66	U26	0.86	UYIN	0.50
6AG7 0.45	6CY7 0.75	68R7 0.50	20D1 0.60	35 D5 0.75	9002 0.45	EC86 0.60	2-20	E¥87	0.48	PCL84 0.46	U31	0.70	UY11	1.00
6AH6 0.60	6D3 0.55	6T8 0.38	20L1 1.10	35L6GT0.75	A2293 2-20	EC88 0.60	EF40 0.50	EY88	0.43	PCL85 0.50	U37	6.60		0.48
6AJ8 0.80	6DC6 0.80	6U4GT 0.70	20P1 0.50	35W4 0.40	AZ1 0.60	EC90 0.35	EF41 0.65	EZ40	0.50	PCL86 0.46	U52	0.40	UY82	0.50
6AK5 0.40	6DK6 0.80	6U5G 0.75	20P4 1.10	35Z3 0.75	AZ31 0.55	EC92 0.45	EF42 0.70	EZ41	0.75	PCL88 1.25	U76	0.40		
6AK6 0.60	6DQ6B 0.75	6U8A 0.48	20P5 1.20	35Z4G 0.40	CBL1 0.99	EC93 0.60	EF80 0.80	EZ80	0.28	PCL800 1.10	U78	0.40	UY85	0.40
6AL3 0.48	6D84 1.25	6V6GT 0.45	25C5 0.60	35Z5GT 0.70	CBL31 1.00	ECC35 1.00	EF85 0.85	EZ81	0.29	PCL801 0.95	U191	0.75	W729	0.75
6AL5 0.22	6EA8 0.65	6X4 0.40	25L6GT 0.50	50A5 0.80	CY31 0.50	ECC40 0.70	EF86 0.30	GT1C	8.00	PD500 1.30	U201	0.60	Y63	0.75
6AM5 0.40	8EH7 0.30	6X5GT 0.45	25Z4G 0.35	50B5 0.70	DAF96 0.50	ECC81 0.40	EF89 0.28	GY501		PF86 0.70	U281	0.55	Z759	8.00
6AM6 0.37	6EJ7 0.35	6X8 0.65	25Z6GT0.75	50C5 0.60	DF96 0.50	ECC82 0.33	EF91 0.87	GZ30	0.45	PF818 1.00	U282	0.55		1.85
6AQ5 0.42		6Y6G 0.80	30 A5 0.60	50CD6G1.20	DK92 0.70	ECC83 0.83	EF92 0.35	GZ31	0.40	PFL200 0.85	U301	0.55	Z803U	1.80
6AQ6 0.70	DE WO 0.70	0.00	1 50 AU 0.00	0002302.00						-				-

TR	Δ	N	S	IST	O	RS	

	TRANSISTORS									
	2N696	0.15	AC128	0.15	BD123	0.80				
	2N697	0.15	AC132	0.35	BD124	0.80				
	2N698	0.80	AC153	0.20	BD131	0.75				
	2N705	0.70	AC154	0.15	BD132	0.80				
	2N706	0.10	AC157 AC169	0.17	BF115 BF167	0.20				
	2N708	0.15	AC176	0.17	BF173	0.20				
	2N753	0.25	AC170	0.25	BF179	0.80				
	2N929 2N930	0.20	AC188	0.25	BF180	0.30				
	2N997	0.80	ACY17	0.25	BF181	0.80				
	2N1131	0.25	ACY18	0.20	BF184	0.20				
	2N1132	0.25	ACY19	0.25	BF185	0.20				
	2N1184	1.25	ACY20	0.20	BF194	0.14				
	2N1302	0.17	ACY21	0.20	BF195	0.14				
	2N1303	0.17	ACY22	0.10	BF196	0.18				
	2N1304	0.20	AD140 AD149	0.50 0.45	BF197 BF200	0.12				
	2N1305	0.20	AD161	0.35	BFW87	0.25				
	2N1306 2N1307	0.25	AD162	0.35	BFW88	0.20				
	2N1307	0.25	ADZ11	1.25	BFW89	0.20				
	2N1309	0.25	ADZ12	1.25	BFW91	0.20				
	2N1613	0.17	AF114	0.17	BFX88	0.20				
	2N1711	0.20	AF115	0.17	BFY17	0.40				
	2N1756	0.75	AF116	0.17	BFY19	0.25				
	2N2147	0.75	AF117	0.17	BFY50	0.20				
	2N2160	0.60	AF118 AF125	0.30 0.20	BFY51	0.20				
	2N2217	0.25 0.20	AF125	0.20	BFY52 BBY26	0.17				
	2N2218 2N2219	0.20	AF180	0.35	BSY27	0.15				
	2N2369A	0.15	AF181	0.35	BS Y 28	0.15 0.17				
	2N2477	0.65	AF186	0.40	BSY65	0.20				
	2N2646	0.40	AF239	0.36	BSY95A	0.12				
	2N2905	0.25	AFZ11	0.45	OC16	0.50				
	2N2923	0.12	AFZ12	1.00	OC22	0.50				
	2N2924	0.13	ASY26	0.25	OC23	0.60				
	2N2926	0.10	ASY27 ASY28	0.25	OC24 OC25	0.85				
	2N3053 2N3054	0.20	A8Y29	0.80	OC26	0.25				
	2N3055	0.60	ASY54	0.25	OC28	0.50				
	2N3133	0.25	ABZ15	0.70	OC29	0.60				
	2N3134	0.25	ASZ16	0.70 0.75 0.75	OC35	0.50				
	2N3391	0.17	ABZ17	0.75	OC36	0.45				
ľ	2N 3392	0.17	A8Z18	0.75	OC42	0.20				
ı	2N3393	0.15	ASZ20 ASZ21	0.25	OC44	0.15 0.12				
ŀ	2N3394	0.15	BC107	0.10	OC45 OC70	0.12				
١	2N3395 2N3402	0.20 0.15	BC108	0.10	0C71	0.12				
ı	2N3403	0.15	BC109	0.10	OC72	0.12				
ı	2N3404	0.32	BC113	0.10	OC73	0.30				
	2N3414	0.20	BC118	0.20	OC75	0.15				
ı	2N3415	0.15	BC134	0.20	OC76	0.15				
ŀ	2N3417	0.25	BC147	0.10	OC78	0.25				
١	2N3702	0.10	BC148 BC149	0.10 0.12	OC78D	0.20				
ł	2N3703	0.10	BC152	0.15	OC81 OC81D	0.20				
l	2N3704 2N3707	0.12	BC158	0.15	OC83	0.20				
ı	2N3707	0.12	BC175	0.20	OC139	0.30				
l	2N3710	0.10	BC186	0.25	OC140	0.35				
ĺ	2N3819	0.35	BCY30	0.25	OC141	0.60				
ĺ	2N3906	0.25	BCY31	0.30	OC170	0.25				
ĺ	28702	0.50	BCY33	0.25	OC171	0.25				
Ì	28746	0.25	BCY34	0.30	OC200	0.30				
ĺ	AC113	0.15	BCY72	0.15	OC201	0.60				
۱	AC125 AC126	0.30	BCZ10 BCZ11	0.30	OC202 OC203	0.40				
į	AC126	0.20	BD121	0.40	OC204	0.40				
ĕ,	AUIZ.	0.17	DULL	2.00	0 0001					







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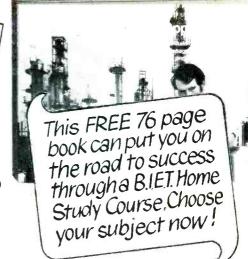
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