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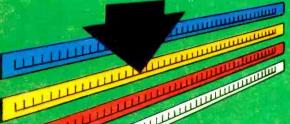


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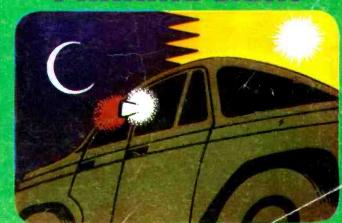
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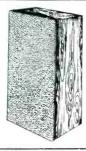
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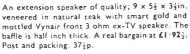
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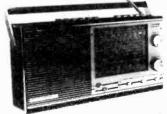
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BP9 4 IN4607 Sit. Rec. diodes. 1,000 PIV lamp plastic. 50p live divided by weight 1,000 PIV lamp plastic. 50p live divided by weight 1,000 PIV lamp plastic. 50p live and values. 50p live and live		TESTE	D AND GUARANTEED PAK	
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## 100 Page and small Page Page	B 79	4		50p
Augustity, counted by weight	B81	10		50p
Harmonia	899	200	Mixed Capacitors. Approx. quantity, counted by weight	50p
## ## ## ## ## ## ## ## ## ## ## ## ##	H4	250		50p
H9 2 OCP?I Light Sensitive H9 2 OCP?I Light Sensitive Flore Transistor H12 50 NKT155/259 Germ diodes, brand new stock clearance H28 20 OC20/11/2/3 PNP Sillicon H39 20 I Watt Zener Diodes, Mixed Voltages 6.6 – 43v. H35 100 Mixed Diodes, Germ. Gold bonded, etc. Marked and Unmarked. H38 30 Short lead Transistors, NPN Sillicon Planar types. UNMARKED UNTESTED PACKS B66 150 Germanium Diodes H38 200 Trans. manufacturers' rejects of the company	Н7	40	Wirewound Resistors, Mixed types and values.	50p
H12 50 NKT155/259 Germ. diodes.	H8	4	BY127 Sil. Recs. 1000 PIV, I amp. plastic	50p
H28 20 OC200/1/2/3 PNP Silicon uncoded TO-5 can	H9	2	OCP71 Light Sensitive Photo Transistor	50p
100 100	H12	50	NKT155/259 Germ. diodes, brand new stock clearance	50p
H35 100 Mixed Diodes, Germ, Gold bonded, etc. Marked and Unmarked. Sop U	H28	20		50p
Unmarked. H38 30 Short lead Transistors, NPN S0p UNMARKED UNTESTED PACKS B66 150 Germanium Diodes Min. glass type B83 200 Trans. manufacturers' rejects of all types NPN, PNP, Sil. and Germ. B84 100 Silicon Diodes DO-7 glass equiv. to OA200, OA202 50p B86 50 Sil. Diodes sub. min. N91 and IN916 types B88 50 Sil. Trans. NPN, PNP equiv. to OC200/1 2N706A, BS795A, etc. B1 50 Germanium Transistors PNP, AF and RF H6 40 250mW, Zener Diodes DO-7 Min. Glass Type H17 20 3 amp. Silicon Stud Rectifiers, 50p mixed voits H18 30 Top Hat Silicon Rectifiers. 50p H19 30 Top Hat Silicon Rectifiers. 50p H16 8 Experimenters' Pak of Integrated Circuits. Data supplied. H20 20 BY126/Type Silicon Rectifiers 50p BY126/Type Silicon Rectifiers 50p H20 20 BY126/Type Silicon Rectifiers 50p	H30	20		50p
B86 100 Germanium Diodes Min. glass type B87 200 Trans. manufacturers' rejects all types RPN, PNP, Sil. and Germ. B88 100 Silicon Diodes DO-7 glass equiv. to OA200, OA202 B88 50 Sil. Diodes sub. min. Silicon Co200/1 Silicon Stub. min. Sup. PNP Sup. April Sup. Sup. Sup. Sup. Sup. Sup. Sup. Sup.	H 35	100	Mixed Diodes, Germ. Gold bonded, etc. Marked and Unmarked.	50p
B86	H38	30	Short lead Transistors, NPN Silicon Planar types.	50p
B83 200 Trans. manufacturers' rejects alt types NPN, PNP, Sil. and Germ.		UNP	ARKED UNTESTED PACKS	
all types NPN, PNP, Stl. and Germ. B84 100 Silicon Diodes DO-7 glass cquiv. to OA200, OA202 50P equiv. to OA200, OA202 50P of the policy of t	B66	150	Germanium Diodes Min. glass type	50p
100 equiv. to OA200, OA202 50p	B83	200	Trans. manufacturers' rejects all types NPN, PNP, Sil. and	50p
Solit Trans. NPN. PNP equiv. to OC200/1 Solp	B84	100	Silicon Diodes DO-7 glass equiv. to OA200, OA202	50p
81 50 Germanium Transistors 50p	B86	50	IN914 and IN916 types	50p
H6 40 250mW. Zener Diodes DOP Min. Glass Type H17 20 3 amp. Silicon Stud Rectifiers, 50p mixed volts H18 30 Top Hst Silicon Rectifiers, 50p T50mA. Mixed volts H16 8 Experimenters' Pak of Integrated Circuits. Data supplied H20 20 BY126/7 Type Silicon Rectifiers 50p I amp plastic. Mixed volts	B88	50	Sil, Trans. NPN, PNP equiv. to OC200/I 2N706A, BSY95A, etc.	
H17 20 3 amp. Silicon Stud Rectifiers, 50p mixed volts H15 30 Top Hat Silicon Rectifiers. 50p 750mA. Mixed volts H16 8 Experimenters' Pak of Integrated Circuits. Data supplied. H20 20 BY126/7 Type Silicon Rectifiers 50p 1 amp plastic. Mixed volts.	ВІ	50	PNP, AF and RF	50p
H15 30 Top Hat Silicon Rectifiers. 50p T750mA. Mixed volts H16 8 Experimenters' Pak of Integrated Circuits. Data supplied. H20 20 BY126/7 Type Silicon Rectifiers 50p T amp plastic. Mixed volts.	H6	40	250mW. Zener Diodes DO-7 Min. Glass Type	50p
H16 8 Experimenters' Pak of Integrated Circuits. Data supplied. H20 20 BY126/7 Type Silicon Rectifiers 50p I amp plastic. Mixed volts.	H 17	20		50p
H20 20 BY126/7 Type Silicon Rectifiers 50p amp plastic, Mixed volts.	H 15	30	Top Hat Silicon Rectifiers, 750mA. Mixed volts	<u>.</u>
l amp plastic. Mixed voits.	H16	8	Integrated Circuits. Data supplied	
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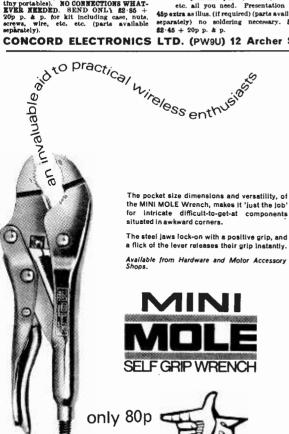
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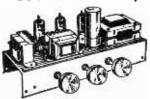


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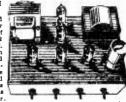
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Response 30-20,000 c.p.s.
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Impedance 15 ohms

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PRINTED CIRCUITRY TWENTY SILICON TRANSISTORS. FOUR DIODES, FOUR RECTIFIERS

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Cabinets latest style Satin Teak veneer. Acoustically lined or filled acoustic damping. Ported where appropriate, Credit terms available.

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c.p.s. Rating 8-10 watts. Fitted High flux 13 × 8in.

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STANWAY 11 Size 20 × 101 × 9in. approx. Rating 10 watts. Inc. 13 × 8in. speaker with highly flexible cone aurround, long throw voice coil and 10,000 line nagnet. High flux tweeter. Handsome Scandinavian design cabinet. Range 35-20,000 c.p.s. Imp. Gives smooth realistic sound output. See 'package offers' for £17.85

R.S.C. TAI2 MKIII 6.5 + 6.5 WATT STEREO AMPLIFIER

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HIGH FIDELITY OUTPUT OF 6.5 WATTS PER CHANNEL
Designed for optimum performance with any crystal
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TYPE BMI. An all-dry battery eliminator. Size 5½×4½×2lin. approx. Completely replaces batteries supplying 1.5v and 90v. to battery radio where A.C. mains 200/250v. Soles is available. COMPLETE KIT 43-25 ASSEMBLED READY 43-75 WITH DIAGRAM 43-25 FOR USE

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* PUSH-BUTTON SELECTOR SWITCHING

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USE WITH THE BEST ANCILLARY EQUIPMENT THAT CAN BE AFFORDED.

OUTPUT: 15 watts R.M.S. (Continuous) into 8 ohms.
HUM & NOISE —75dB Min. Vol. —65dB Full Vol. HARMONIC DISTORTION
FREQUENCY RESPONSE: —3dB 7Hz to 70MHz
TRBBLE CONTROL: +16dB to —12dB at 14Hz
BASS CONTROL: +17dB to —16dB at 40Hz
SENSITUTITIES: Disc Mag. 2:5mV. Ceramic 35mV. Redio 120mV. Tape 120mV,
REAR PANEL SOCKETS ARE FOR 3 PAIRS OF INPUTS (1) P.U. (2) Radio.
(3) Tape Amp. Plus pair for tape recorder signal take off and 2 pairs for speaker



R.S.C. COLUMN SPEAKERS IDEAL FOR VOCALISTS

All types 15 Ohms covered in Revine and Vynair
TYPE C4100 IS ALSO SUITABLE FOR BASS GUITAR OR ELECTRONIC ORGAN

TYPE C482 25-30 WATTS
Fitted four 8° high flux 8 watt speakers
Overall size approx. 48 x 10 x 5in. £17-75
Terms: Dep £3 and 9 mouthly 2
payments £2 (Total £21)

Carr. 50p.

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TERMS: Deposit 24 and 9 monthly payments of 22.10 (Total 222.90). Send S.A.E. for leaflet.

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250-0-250v. 100mA, 6·3v. 2a., 0·5-6·3v. 2a. 21·35
250-0-250v. 100mA, 6·3v. 4a., 0·5-6·3v. 3a. 22·20
300-0-300v. 100mA, 6·3v. 4a., 0·5-6·3v. 3a. 22·20
300-0-300v. 130mA, 6·3v. 4a., 0·5-6·3v. 1a. 21·30
350-0-350v. 100mA, 6·3v. 4a., 0·5-6·3v. 3a. 22·20
350-0-350v. 100mA, 6·3v. 4a., 0·5-6·3v. 3a. 22·20

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Black Rexine covered. Size approx.
18 × 18 × 8 ins.

Terms on Amps. Speakers and Head-phone/Mike. Deposit £15 and 18 monthly payments of £3.95 (Total 286-10).

F.G.1/2A DISCO CONSOLE Incorporating twin Garrard SP25
Mk.II turntables and Ceramic
Cartridges with diamond stylii. Mk.III urntables and Ceramic Cartridges with diamond stylii. Separate Vol. controls for each turntable. Also MONITORING FACILITIES, plus Treble and Bass Controls. Separate input for 'mike' with vol. control switch. Black Rexine covered Cabinet with lid. see illus. on left Carr. £1.25 £67.10

Or Dep. £13.25 and \$6.75 (Total £74) or Dep. £15 and 18 mithly paymis £5.56 (Total £74) or Dep. £15 and 18 mithly paymis £5.56 (Total £79.68).

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£12.90

All power ratings are R.M.S. continuous 2 TEARS GUARANTEE
High flux ceramic magnets. ALL CARR. FREE
POP' 100 'POP' 60 'POP' 50 15" 60 Watt 14,000 gauss 8/15Ω

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£22.95 Dep: 26 and 9 mthly payments \$2.20 (Total \$25.80)

Dep. 23-30 and 9 monthly payments 21-30 (Total 215). FOR BASS GUITAR, ELECT. ORGAN, ETC.

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Dep 22 and 9 monthly pay-ments 21.20 ments £1.20 (Total £12.80) PAIR SUITABLE ALL PURPOSES

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Dual Cone 15 Ω (for uses £6.75 Carr. or Dep. £1 and 9 mthly other than Bass Guitar or Free Electronic Organ).

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PACKAGE PRICE £84-95 £21-80 £52 CAFT £56.75

PACKAGE PRICE £84.95 £23.90 £56 carr £1.25 458-85

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ALL TWO TONE REXINE AND VYNAIR FINISH L125 50 WATT Fitted pair of 12" 50 watt high flux speakers for conservative rating. Impedance 8-15 ohms. Carr. £1-50 or deposit £4-50 and 9 nonthly pays. of £3-62. Total £37-08

Or deposit \$4.50 and 9 nonthly payes of \$2.52. Note 2015

L12/25 | 12" 25 WATT | L13 13" × 8" 10 Watt |
10,000 lines 3 or 15 ohms, State impedance required. Carr 40p £5:25

£61.95 £48-60

£105-55

Carr. 50p

FAL PHASE 50 MkIII AMPLIFIER 50W

RSC BASS-REGENT **50 WATT AMPLIFIER**

A powerful high quality all-purpose unit for lead, rhythm, bass guitar, vocalists, gram, radio, tape. Peak Output rating Loudspeaker unit Output rating, optional horiz horizontal

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† Two extra heavy duty 12in. 50w L'spkrs.

Four Jack inputs and two Volume
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For Guitar, Vocal or Instrumental Groub.

4 input, 2 vol. control Hi-Fi 30 watt unit with Separate Bass and Treble controls. Current valves. Peak output rating, Strong Rexine covered cabinet with handles. Attractive black/gold P.V.C. facis. Neon indicator. For 200-250v. A.C. mains. For 3 or 15 ohm speakers. Send S.A.E. for leaflet. Terms: Deposit 24:30 and 9 monthly payments of \$2:30 (Total \$25:00).



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0-200-230-250v. 50 Hz A.C. mains operation. Inputs for magnetic or Ceramic Pickup, Tape or Radio Tuner.

6+6 WATTS

Recommended Retail Price

TECHNICAL DETAILS

Bass Control ± 12 dB at 40 Hz. Treble Control ± 12 dB at 14 KHz

Treble Control ± 12 up at 19 kmz.

Sensitivities Mag. P.U. 3.5 m.v.
Into 47K ohm R.I.A.A. Ceramic
P.U. 35 m.v. into 100K ohm. Tape
Amp. 100 m.v. into 100K. Radio
Tuner 400 m.v. into 400K ohm, Crosstalk 53 dB

Hum and Noise-75 dB min. vol. -65 dB max. vol.

Total Harmonic Distortion 0 1% at 1watt into 15 ohms.

Output (per channel) 6-5 watts I.H.F.M.

★ Individual Bass and Treble Controls.

Frequency Response ± 1½ dB 20 Hz to 65 KHz.

Outputs for Speaker impedances between 3 and 15 ohms.

* Stereo/Mono Switch.

★ Input Selector Switch.

* Solid State Circuitry.

* Attractive silver finished metal facia and matching control knobs.

- * A modestly priced solid state unit.
- * The Silver Facia with black lettering enhanced by matching control knobs, provides a high standard of appearance
- * Suitable for crystal Gram. Pick-up cartridges and Radio input.
- * A wide range of tone variation is provided by the separate Bass and Treble 'lift' and 'cut' controls.
- A selector switch permits instantaneous selection of Gram. or Radio.
- ★ Speaker impedances between 3 and 15 ohms are suitable.

Wholesale and Retail enquiries to the Manufacturers

TECHNICAL DETAILS

Frequency Range 20 Hz to

Output (per channel) 5 watts

Bass Control ± 12 dB at 60 Hz. Treble Control ± 14 dB. at 14 KHz.

PRINTED CIRCUIT CONSTRUCTION EMPLOYING 10 TRANSISTORS



0-200-250v. 50 Hz A.C. mains operation

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EB91	(0.09)	PCC189	0.47	PY801	0.32
ECC81	8:15	PCF80	0.27	UABC80	0.30
ECC82	185	PCF86	0.44	UCC85	0.34
ECC83	0.24	PCF802	0.38	UCH81	0.30
ECL80	0.25	PCF806	0.56	UCL82	0.31
EF80	0.22	PCF808	0.68	UL41	0.54
EF85	0.26	PCL82	0.30	UY85	0.24
EF86	0.28	PCL83	0.55	6/30L2	0.58
EF89	0.23	PCL84	0.33	6AT6	0,18
EF91	(a.J2)	PCL85	0.37	6AU6	0.20
EF183	0.26	PCL86	0.37	6BA6	0.20
EF184	0.28	PFL200	0.51	6BE6	0.21
EH90	0.35	PL36	0.47	6F23	0.67
EL41	0.53	PL81	0.43	I2AT6	0.23
EL84	0.22	PL82	0.29	I2AT7	0.15
EY51	0.30	PL83	0.31	12AU7	0.19
EY86	0.28	PL84	0.29	I2AX7	0.24
EZ80	9.20	PL500	0.61	I2BA6	0.30
EZ81	7 0.21	PL504	0.61	12BE6	0.30
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Manual £8 50 Twin Speaker £8-95 Push Button £10 45

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Quality Cassettes at cheapest prices C60 35p plus 5p P. & P. C90 47p plus 5p P. & P. C120 60p plus 5p P. & P.

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Radio AM with Battery, Magnetic Earphone and Carrying Case, £1 95 Postage Paid.

Mixed packet of resistors 25p postage paid

Build yourselfa TRANSISTOR RADIO

NEW! ROAMER 10 WITH VHF INCLUDING AIRCRAFT

10 TRANSISTORS. 9 TUNABLE WAVEBANDS. MW1. MW2, LW, SW1, SW2, SW3, TRAWLER BAND. VHF AND LOCAL STATIONS AND AIRCRAFT BAND

Built in Perrite Rod Aerial for MW,LW. Retractable, chrome plated Telescopic Aerial, for peak short wave and VHF listening. Push Pull output using 690mw Transistors. Car Aerial and Tape Record Sockets. Switched Earpiece Socket complete with Earpiece 10 Transistors plus 3 Diodes. 8° x 24° Speaker. Air Spaced ganged Tuning Condenser with VHF section. Volume onjoff, Wave Change and Tone Control. Attractive Case in black with silver blocking. Size 9° x 7° x 4°. Easy to follow instructions and diagrams. Parts price list and easy build plans 30p (FREE with parts).

Total building cost

£8·50

(Oversens P & P. El)



ROAMER FIGHT Mk I NOW WITH VARIABLE TONE CONTROL

7 Tunable Wavebands: MW1, MW2, LW, SW1, SW2, SW3 and Trawler Band. Built in Ferrite Rod Aerial for MW and LW. Retractable chrome plated Telescopic aerial for Short Waves. Push pull output using 600mW transistors. Car aerial and Tape record sockets. Selectivity switch. Switched earpiece socket complete with earpiece. 8 transistors plus 3 diodes. 8° × 2° Bpeaker. Air spaced ganged tuning condenser. Volume/on/off, tuning, wave change and tone controlled Attractive case in rich chestnut shade with gold blocking. Size 9 × 7 × 4 in. approx. Easy to follow instructions and diagrams. Parts Price List and Easy Build Plans 25p (FREE with parts).

Total building cost 26-98 P. P. & Overseas P. & P. £1)

ROAMER SEVEN MK IV

7 Tunable Wave-banis: MW1, MW2, LW, SW1, SW2, SW3 and Trawler Bard. Extra Medium

BW3 and Trawler
Bard. Extra Medium waveband provides easier tuning
of Radio Luxembourg, etc. Built in ferrite rod aerial
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Povertin jush-pull output, 7 translators and 2 diodes,
8° x 24' speaker. Air spaced ganged tuning condenser.
Volume/onlor, tuning and
wave change controls.
Attractive case say to follow instructions and diagrams. Parts
price list and easy build plans 150 (FREE with parts).
Easpiece. Basy build plans 150 (FREE with parts).
Easpiece with plug and switched socket for private
listening, 800 extra.

Total building costs £5-98 P. P.&

ROAMER

6 Tunable Wave-bands: MW, LW, SW1, SW2, Trawler band plus an extra M.W. band for easier tuning of Luxembourg etc. Sensitive fer-rite rod aerial and telescopic aerial



for Short Waves. Sin. Speaker. 8 stages—6 transistors and 2 glodes Attractive black case with red grille, dial and black knobs with polished metal inserts. Size $9 \times 51 \times 21$ in. approx. Easy build plans and parts price list 15p (PREE with parts). Earpiece with plug and switched socket for private listening 30p extra.

Total building costs £3-98 P. P. &

POCKET FIVE

3 Tunable Wavebands: MW, LW, Trawler Band with extended M.W. band for easier tuning band for easier tuning of Luxembourg, etc. 7 stages—5 transistors and 2 diodes, supersensitive ferrite rod serial, fine tone moving coil speaker. Attractive black and gold case. Size 6 × 1½ × 34 in. Easy build plans and parts price list 10p (FREE with parts). Earpiece with plug and switched socket for private listening 30p

Total building costs £2-29 P.P. & Ins. 21p

TRANSONA FIVE

5 TRANSISTORS AND 2 DIODES



3 Tunable Wavebands: MW, LW and Trawler Band. 7 stage—5 transistors and 2 diodes, ferrite rod aerial. tunning condenser volume control, fine tone moving coil speaker. Attractive case with red speaker griffe. Size 6½ × 4½ × 1½ in. Easy build plans and parts price list 10g (FREE with parts). Earpiece with plug and witched socket for miviate listening 30m extra socket for private listening 30p extra.

Total building costs £2.50 P.P. &

TRANS

8 TRANSISTORS and 3 DIODES

6 Tunable Wave-bands: MW, LW, SW1, SW2, SW3 and Trawler Band. Sensitive ferrite rod aerial for M.W. and L.W. Telescopic aerial for Short Waves. 3ln. Speaker. 8 improved type transistors plus 3 diodes. Attractive case in black with red grille, dini and black knobs with polished metal innerts. Size 9 x 5½ x 2½ in. approx. Push pull output. Battery economiser switch for extended battery life. Ample power to drive a larger speaker. Parts price list and casy build plans 25p (FREE with parts). Earpiece with plug and switched socket for private listening 30p extra.

Total building costs £4.48 P. P. &

EIGHT

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UNITS INCLUDING MASTER UNIT TO
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POCKET FIVE		EDU-KI	Т	
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DRILL CONTROLLER NEW IKW MODEL Electronically changes speed from mately 10 revs. to maximum. Full power at a speeds by finger-tip ontrol. Kit includes all parts. case, everything and full instructions. 21-50 plus 13p post and insurance, made up model also available. 22-25 plus 13p post & p.

MAINS OPERATED CONTACTOR
220/240v. 50 cycle solenoid
with laminated core so very
silent in operation. Closes 4
circuits each rated at 10 amps. Extremely well made by a German Electrical Company. Overall size $2\frac{1}{2} \times 2 \times 2$ in.



Nr. -4 D-1--

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CAR AFRIAL

with dashboard control switch-fully extendable to 40in. or fully retractable. Sultable for 12v positive or negative earth. Supplied complete with fitting instructions and ready wired dashboard switch. \$5.75 plus 25p post and ins.



MAINS TRANSISTOR POWER PACK

PACK
Designed to operate transistor sets and amplifiers. Adjustable output 6v., 9v., 12 volts for up to 500mA (class B working). Takes the place of any of the following batteries: PP1, PP3, PP4, PP6, PP7, PP9, and others. Kit comprises: mains transformer rectifier, smoothing and load resistor condensers and instructions. Real snip at only #1 plus 18p postage.

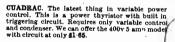


amp changeover contacts, 9p sch, 21 doz. 15 amp Model 0p each or \$1.05 doz.



MINIATURE WAFER SWITCHES

WAFER SWITCHES
2 pole, 2 way—4 pole, 2 way—
3 pole, 3 way—4 pole, 3 way—2
pole, 4 way—3 pole, 4 way—2 pole
6 way—1 pole, 12 way. All at 20p
each, £1.80 for ten, your assortment



FLEX CABLE SNIP

3-core heavy circular T.R.S. waterproof flex, ideal for running down the garden to pool or shed. 1-5mm. cores (5 amp). 100 yard coils \$4.25 plus carriage: 75p up to 200 miles; \$1 300 miles; **21.50** 500 m



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12 WAY SUB-MINIATURE
MULTI-CORE CABLE
7-0076 copper cores each core P.V.C. insulated and
of different colour. P.V.C. covered overall and
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2 square full vision for flush mounting. Moving iron instrument. Id ... charger. Price 60p each. 10 for 15-46.



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Almost zero resistant in sunlight increases to 10 K Ohms in dark or dull light, epoxy
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Rated at 500 MW, wire ended. 489 with circuit.
Also ORP 12 light cell 459. Rated as pow man and a state of the state of



Complete Kit (except wooden battens) to make the metal detector similar to the circuit in Practical Wireless August issue. \$3.95 plus 20p post and insurance.

BAKELITE INSTRUMENT CASE

Size approx. 61" × 31" × 2 deep with brass inserts in four corners and bakelite panel. This is a very strong case suitable to house instruments and special rigs, etc. Price



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This is a fully screened intermediate frequency module for amplification and detection of f.m. signals at 10.7MHz and a.m. signals at 470kHz. The first stage is used as an if. amplifier for f.m. and a self-oscillating mixer for a.m. operation, in conjunction with an external oscillator coil. 75p each. 10 for \$8.75. 100 for \$82.50p. With connection dig.

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Standard size 14" wafer—silver-plated 5-amp contact, standard are ry water—shver-placed p-amp contact,

140. Of Toles	~ way	3 way	* way	o way	o way	o way	9 way	10way	12 way
1 pole	40p	40p	40p	40p	40p	40p	40p	40p	40p
2 poles	40p	40p	40p	40p	40p	40p	40p	70p \	700
3 poles	40p	40p	40p	40p	70p	70p	70p	95p	950
4 poles	40p	40p	40p	70p	70p	70p	70p	#1 -20	£ \ 20
5 poles	40p	40p	70p	70p	95p	95p	95p		\$1 45
6 poles	40p	70p	70p	70p	95p	95p	95p		£1 ·70
7 poles	70p	70p	70p	95p	£1 ·20	£1 ·20		£1 95	£1 95
8 poles	70p	70p	70p	95p	£1 ·20	£1 -20	£1.20	£2.20	£2 ·20
9 poles	70p	70p	95 p	95p	£1 45	£1 ·45	£1 ·45	£2·45	£2.45
10 poles	70p	70p	95p	£1 ·20	£1 ·45	£1 ·45	£1 ·45		£2·70
11 poles	70p	95p	95p	£1 ·20	£1 ·70	£1 ·70	£1 .70	22.95	42.95
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13 AMP TWIN GANG SOCKETS

offered at less than wholesale price your opportunity to replace those dangerous adaptors—brown bakelite flush mounting—standard fitting. Unswitched 20p each, separately switched 30p each. Separately switched and with neon on/off indicators 45p each. Less 10% ten or more +20p postage if order

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YOUR TIME is the most precious thing you have. Do you waste it waiting for the soldering iron to heat. Do you waste it waiting for the soldering iron to heat. Do you can be soldering in a few seconding to the soldering in a few second in the soldering in a few second in the soldering in a few second light yeen properties in fact this month's snip. A well made lightweight untit with flash lamp to illuminate the work. Has 100 wait with flash lamp to illuminate the work. Has 100 wait obuble insulated mains Transformer and is built into a shockproof Thermo-plastic case. Sultable for 240volt, 50 c.p.s. This comes complete with 6 spare tips and is offered at a special snip price \$2.25. plus 20p post and insurance.



CENTRIFUGAL FAN

Mains operated, turbo blower type. Pressed steel housing contains motor and impeller. Motor is 1/10th h.p. giving considerable air flow but virtually no noise. Approx. dimensions 104 in. wide × 12 ins. dia. outlet into trunking 10½ × 4½ in. £496 plus £1 post and insurance.

in. 24-95 plus £1 post and insurance.

THERMOSTAT WITH THERMOMETER

Made by Honeywell for normal air temperatures 40°-80°F.
(5-28°C). This is a precision instrument with a differential
which can be adjusted to better than 1-5°F. A mercury
switch breaks on temp. rise—the switch is operated by
colled bi-metal element and an adjustable heater is incorporated for heat anticipation. Elegan by tyled and encased in an ivory plastic case with clear positic windows,
the transmeter shows and switch ections scale below. Size
enjrox. 38° × 32° × 14° deep. Lan by
conduit box or directly on wall: Tice £1.25 cash or 10 for
£11.25.



Uses 4 transistors, and has an output of 500mW into 8 unins speakers. Input suitable for crystal mic. or pick-int. 9 volt battery operated. Size 2 long × 14" with × 1" limb SPECIAL SNIP PRICE 609 each. 11 or 25.00 214 A A



Just, hat you need for work bush carts: 4 × 13 amp sockets in metal box to take a many 1 13 mm and the plugs and on/off switch with neon warning light. Supplied complete with 7 feet of heavy cable. Wired up ready to work, 22.25 less plug; 22.50 with fitted 13 amp plug; 22.65 with fitted 15 amp plug; plus 23p P. & P.

I HOUR MINUTE TIMER

....... 20.0

Made by Smiths. Complete with control knob and calibrated dial. This month's special bargain at 50p. Useful in the kitchen.



ELECTRIC TIME SWITCH

Made by 8miths these are A.C. mains operated. NOT CLOCKWORK. Ideal for mounting on rack or shelf or can be built into box with 13A socket. 2 completely adjustable time periods per 24 hours, 5 amp change-over contacts will switch circuit on or off during these periods 22.50 post and ins., 23p. Additional time contacts 80p pair. 3000 WATT IMMERSION HEATER

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50 MICRO AMPMETER

Square, panel mounting type: 52 MAINS OPERATED



Model 400/1 I* pul Size 21" × 2 × 11"

75p Model TT10 11" pull size $3 \times 21 \times 2$ £1.80 plus 20p. post and insurance.

MAINS RELAY BARGAIN

Special this month are some single, double and treble pole changeand trenie poje change-over relays. Contacts rated at 15 amps. Operating soil wound for 240V A.C. Good Brittsh Make. Unused. Size approx. 1;" x 1". Open construc-



Single pole 25p each 10 for 22:25 Treble pole 35p each 10 for 28:15

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This system which has proved to be amazingly efficient. We offer kit of parts as P.W. Circuit \$5.95 plus 20p p. & p. Delaxe model with premared circuit board \$6.95. When ordering please state—whatter for positive or negative systems.

or negative systems.

22 POSITION SOLENOID OPERATED STUD

SWITCH
Mains operated, each current pulse to switch
solenoid moves switch arm through one position
on to the next contact stud—current to release
solenoid brings back switch arm to position one.
These are ex-equipment but in good working
order. Any not so would be replaced. Price 509

order. Any not so would be replaced. Fine way each.

TIMED 'ON' SWITCHES

Made by SWITCHES

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It is turned. A full 360° turn or only a part turn winds the clockwork mechanism and keeps the switch closed until the spindle returns to the 'off' position. A dial therefore could be fitted to indicate hours and minutes and the switch closed until the returns to the order of the switch closed until the spindle returns to the 'off' position. A dial therefore could be fitted to indicate hours and minutes and the switch on period could be set quite accurately. 3 models available—90 minutes. —120 minutes and 360 minutes. Price \$50 each. Metal clad pointer knob 15ps Suitable relay to make the switch 'on' lastead of 'off' 35p.

MOTOR GRENERATOR

Ex Admiralty—24 volt D.C. input—240v 50 cps

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MOTOR GEMERATOR

Ex Admiralty—24 volt D.C. input—240v 50 cpa output. Admiralty rating 80 watt but we have tested this to 50% overload voltage regulated so suitable to operate TV or instrument. In case with metal cover—controls on front include voltmeter. Probably cost £200 each to make. Our price only £25 each plus carriage. £2 up to 200 miles £4 up \$\cdot 4.00 miles n mile

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but two of these work ideally together off our standard 240 voit mains. A really beautiful motor, extremely quiet running and reversible. \$1.50 each. Postage one 23p,



two 33p. 230v model #3. PRESSURE SWITCH

PRESSURE SWITCH
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TAPE HEADS
Ministure size, approximately 1 square by 1 long

Miniature size, approximately | square by | long.
These are double wound for stereo. Offered at a
very low price of 45p eaco. New and unused

very low price of 459 eaco. New and unused LEYER SWITCH REF. H52/4

Idea for intercom or similar. Pressing the lever down operates 6 pairs of change-over contacts, pressing the lever up operates 4 pairs of change-over contacts. The switch is spring loaded and normally returns to the off or centre position. Size approximately 1\(\frac{1}{2}\) long x 2\(\frac{1}{2}\) deep x \(\frac{1}{2}\)* thick.

40p each.

PUSH BUTTON SWITCHES

Mains, suitable for audio or R.F. Each switch
rated at 250v 15 amps. 1st (black push button) operates
one change over. 3rd (white push button) operates
one change over. 3rd (white push button) operates
one change over. 4th (white push button) operate
one circuit. Note: all depressed buttons remain
down until cleared by the 5th (red button).
Further note:—It is a relatively easy job to
alter the position of the tags thus making the
switches suit your circuit. Fitted with 3 white
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52.70.

7 POS ROTARY MAINS SWITCH

Rated 15 amp at 230v. 4 circuits. Position A—all circuits open. Position B—circuit 1 closed. Position C—circuit closed. Position C—circuit closed. Position D.—circuit at 2, 3 & 4 closed. Position E—2 & 3 circuits closed. Position F—1, 2, 3, & 4 circuits closed. Position G—2 & 4 circuits closed. 15p each or 10 for £1.35

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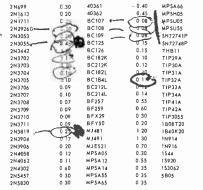
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ı	U5	60 200m A sub-min. Sil. diodes	. (705
1	U6	30 Silicon planar transistors NPN sim. BSY95A, 2N706	. 0.5
1	U7	16 Silicon rectifiers Top-Hat 750mA up to 1,000V	
١	U8	50 Sil. planar diodes 250mA, OA/200/202	
١	U9	20 Mixed volts 1 watt Zener dlodes	
ı	Ull	30 PNP silicon planar transistors TO-5 sim. 2N1132	
١	U13	25 PNP-NPN sil. translators OC200 & 28104	
ı	U14	150 Mixed silicon and germanium diodes	
ı	U15	25 NPN Silicon planar transistors TO-5 sim. 2N697	
٦	U16	10 3-Amp silicon rectifiers stud type up to 1000 PIV	
1	U17	30 Germanium PNP AF transistors TO-5 like ACY 17-22	
ı	Uis	8 6-Amp silicon rectifiers BYZ13 type up to 600 PIV	
Į	U19	25 Silicon NPN transistors like BC108	
1	U20	12 1.5 Amp silicon rectifiers Top-Hat up to 1,000 PIV	
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Į	U23	30 Madt's like MAT series PNP transistors	
ł	U24	20 Germanium 1-Amp rectifiers GJM up to 300 PIV	
ı	U25	25 300 Mc/s NPN sillcon transistors 2N708, BSY27	
Į	U26		0.50
ı	U29	10 1-Amp 8CR's TO-5 can up to 600 PIV CRS1/25-600	
1	U31	20 Sil. Planar NPN trans. low noise amp 8N3707	-
1	U32	25 Zener diodes 400mW D07 case mixed volts, 3-18	
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BP935	13p	12p	11p
BP936	13p	12p	11p
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R P946	12p	11p	10p
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BP951	65p	60p	5 5 p
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Cathode Current (mA)	2.3	14	8	All indicator
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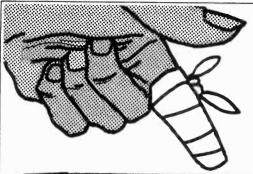
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FERGUSON 3248 4-track FERGUSON 3252 4-track FERGUSON 3258 4-track FERGUSON 3258 4-track FERROGRAPH 702/4 Dolby FERROGRAPH 702/4 Dolby FERROGRAPH 722/4 Dolby FERROGRAPH 722/4 Track tape deck	53.75 100.52 72.40 264.38 274.95 301.98 312.55 224.43	74 · 95 53 · 95 224 · 50 233 · 95 255 · 95 264 · 95 170 · 65
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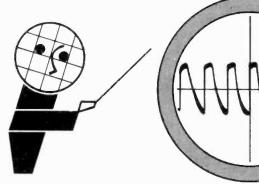
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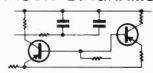
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RAPY

WIRELESS

VOL 48 NO 5

Issue 787

SEPTEMBER 1972

Anyone for Going Back?

ON looking at our front cover this month, readers may be forgiven for thinking that due to shortage of material we are having to ransack the archives! We hope the following words of explanation will dispel any such unworthy and infamous notions.

First of all, the response to our Going Back feature is large enough to indicate a real interest in the adolescence of electronics. Perhaps some people are becoming blase about modern technological progress; perhaps the erstwhile mystery and romance has fled the scene; perhaps the pioneer days are now sufficiently remote in time to give an aura of the unknown; perhaps—but why go on?

Secondly, wireless, radio, electronics (call it what you will) is in danger of becoming taken far too seriously by the amateur, when it should be a relaxation. It should be

fun.

We recently visited the Southampton plant of Mullard and saw how the integrated circuit designer, having determined the basic performance requirements and circuit elements, then enlists the aid of a computer with disc stores and a graphics unit. He can simulate circuits on the computer and obtain a readout of predicted performance.

Having found the optimum circuit elements, he then calls on the computer to provide cost and performance comparisons using different technologies of manufacture. The computer then takes the designer's rough IC layout and feeds in his information, using a special 'language' calls out transistor types from the computer memory store. The drawing for the mask is then generated by the computer.

To check that the circuit is correct, the engineer feeds his circuit programme into the computer, which examines the circuit and its geometry. The engineer then asks the computer to generate a control tape to reproduce the circuit on a photographically-treated glass plate reticle

ten times up on final circuit size.

As a parallel operation, the computer examines the logic design of the circuit and is programmed to determine which test procedures and stimuli are wanted for the computer-testing of the complete circuit. Chips incorporating up to 6,000 transistors and all their interconnections can be designed without making a single error.

For those who boggle at such marvels, we thought it would be therapeutic to dip into the past and publish something from the days when wireless was young and unabashed. Hence the 'reproduction' crystal set. It should even please the transistor men who have been lambasting the valve lovers—after all, it is solidstate!

W. N. STEVENS—Editor

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THE OCTOBER ISSUE WILL BE PUBLISHED ON SEPTEMBER 1st

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NEWS...

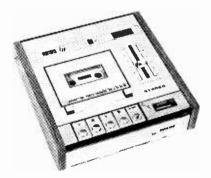


tuner/amplifier/cassette recorder with built-in DNL-the first to include this feature.

The amplifier is rated at 2×10 watts (music) into 4 ohms, with a signal to noise ratio of better than 50dB. Inputs are provided for gram, microphone and external recorder. The tuner in-

cludes long, two medium, short and f.m. wavebands. Decoder and stereo signal lamp are built-in as is a.f.c. Wavebands are selected by recessed push buttons located beneath the fascia. Stations are selected through a large diameter control to the right of the fascia.

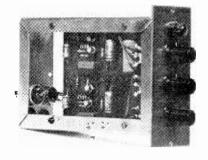
Push buttons on the top of the cabinet control the stereo cassette recorder which is linked internally for trouble-free direct recording in stereo or mono. The cassette unit incorporates automatic recording level, pause con-



trol, cassette eject control and 3 digit counter. The RH811 is finished in walnut and matt chrome and will be supplied with matching RH411 speakers. Retail price is £125.

Also announced is the N2506: a sophisticated high specification mains stereo cassette deck with built in Dynamic Noise Limiter. The DNL is Philips special electronic circuit for reducing the hiss content of a signal—especially noticeable in soft passages of music. The DNL facility is switchable. Recommended retail price £60.

Audio Modules



A new middle of the road range of audio preamplifier and mixer modules have been produced by Partridge Electronics, who until recently specialised in the manufacture of audio mixers to customers own individual require-ments. They have now produced a range of units to provide a flexible and economical method of constructing small or large mixers. The majority of these are available in kit form, or fully assembled and tested. Most items use a standard panel of 16

gauge anodised aluminium, with black lettering, size $50 \times 150 \text{m/m}$. A free fourteen page catalogue is available on receipt of 5p for posting and packing, on application to Partridge Electronics, 21/ 25 Hart Road, Benfleet, Essex.

Figure it out

West Hyde Developments Limited introduce a new, small $7 \times 9^{1}_{2} \times$ 314in. 312lb. calculator which multiplies and divides. The arithmetic can be mixed, i.e. 2+6-3 $\times 5-6 \div 2+4 \cdot 5=14$. The decimal place can be set with 0 to 7 digits



to the right and the machine automatically clears after the equals sign is pressed. Price is £99 for 1 off dropping to 2 at £89 and £75 for 10 off.

Features include 16 digits and a constant factor, 6 months guarantee, Pandicon display tubes (Mullard), 6 LSI chips, 1 I.C., Plug in printed circuit board, Plug in keyboard (gold plated connectors), Tobicons available on sale or return, West Hyde Developments Ltd., Ryefield Crescent, Northwood Hills, Middlesex, HA6 1NN.

Tape Q and A

The second edition of Heinz Ritter's 120-page book, "Tape Questions & Tape Answers" is now available from the Audio/ Video Tapes Division of BASF UK Ltd.

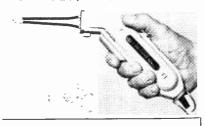
The book contains useful hints and tips on all aspects of tape recording and costs only 40p.

P.O. Box 473, Knightsbridge House, 197 Knightbridge, London, S.W.7.

NEWS...

NEWS...

NEWS...



Amtron kits

Amtron, the Italian electronics company recently sent us some samples of their range of construction kits. As yet, the company have not appointed a sole agent and all enquiries should be made to Box 102, Advertisement Dept., Practical Wireless, Fleetway House, Farringdon Street, London, EC4A 4AD.

The kits are packed in "bubble packs" and contain all the components right down to the last nut and bolt and in many cases a printed circuit board.

The Twilight Switch (shown) circuit contains a photo-electric cell, two transistors, two diodes and can switch up to 1kW using a relay. The unit is activated by a wide range of light intensities and the action is reversible. It can be set to operate at pre-determined light intensities as light falls or increases, making it ideal for switching on a household light at dusk when the owners are out or away on holiday.

Power source is 220-240V a.c. The other unit illustrated is a 27MHz Radio Control Receiver, designated UK 345. This is a superhet employing four highgain, low-noise transistors and one detector diode. The oscillator is crystal controlled. Supply voltage is 6V and current consumption 5mA.



Primaxa gun

S. Kempner Ltd. announce their Primax and Primaxa soldering guns. Full heating is obtained six seconds after the "on" switch is depressed and twin spotlights mounted in the gun enable the user to get "a bit of light on the subject". The special alloy tip never needs retinning and the unit is guaranteed for 12 months (except tips and bulbs). Replacement tips are available for different soldering jobs and two models of gun are available: the 60W model which retails at £5.40 and the 100W model priced at £7. Postage and packing on each solder gun is 25p. Further details and soldering guns are available from S. Kempner Limited, 421 High Road, London, N.12. Tel. 01-346 6222.

Zeta windings

Zeta Windings Ltd. inform us that they can undertake to manufacture and design iron/ferrite cored transformers for any application that constructors may require for the development of their circuitry. They will also make them for P.W. projects and they are available by type number via: Tidman Mail Order, 236 Sandycombe Road, Richmond; H. L. Smith, Edgware Road. 287 London, W.2 or (callers only) Zeta Windings Ltd., 26 All Saints Road, London, W.11.

Other services provided by Zeta Windings are their rewinding of one-off prototype facilities on all kinds of transformers, and their rewind and prototype service which takes about 3 to 5 days.

The "voice"

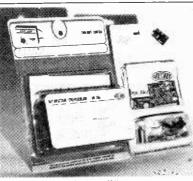
In a recent experiment, Bell Laboratories and the Western Electric Company gave a computer a "voice" so it could speak instructions helpful to productionline workers assembling complex communications equipment.

Previously, computers had been used to calculate efficient wiring instructions, and to print them in typewritten form. To get the instructions into a spoken form—so the assembler need not divert eyes or hands from the job—they were read aloud by a human announcer, and recorded, checked and corrected by technicians. Frequent changes in the complex wire lists meant frequent rerecording and re-checking.

In the recent experiment, the human steps were bypassed, and the computer made to "speak" the wiring instructions directly to an automatic recorder, which were then tested on the production line by an assembler of telephone crossbar switching equipment.

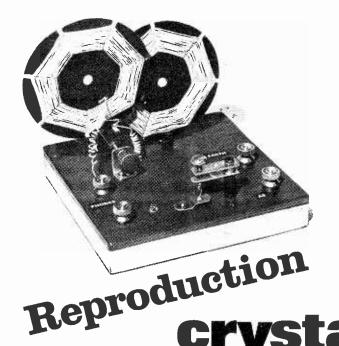
"Black ... three ... thirty-one A ... terminal strip ... four A ... valve socket," the recording said, to "tell" the assembler information about wire colour, length, and connection points. Using a cassette tape player with a foot pedal for control, she could start or stop the recorded voice during any of 58 steps involved. Assembly of each crossbar unit takes about 15 minutes.

Bell Labs engineers have been doing research on computergenerated "speech" because this is one way to make the computer respond in human terms.



Amtron Kits





Next November the British Broadcasting Corporation will be commemorating the first broadcasts, made in 1922, by the original British Broadcasting Company. We thought that readers might be interested in listening to these 50th anniversary programmes on a reproduction crystal set, typical of receivers in use in those early days.

Crystal set R.F.GRAHAM

In the early days of radio crystal sets were used in many homes, and this project is a reproduction of such a receiver. Variable capacitors were available for tuning, but their cost often resulted in some other means of tuning such as tapped coils, coils with sliding contacts bearing on the turns and swinging coils where mutual coupling (and hence the inductance and resonant frequency) could be adjusted.

In addition to the headphones, such receivers required only a few components such as terminals, a detector crystal and "catswhisker", wire to wind the coils, an insulated board or panel and a few small parts such as bolts and brackets to make a detector assembly.

The receiver shown here is something of a novelty, and sure to arouse interest when it is seen.

COILS

These are a flat type quite popular in the early days, and wound with 26 s.w.g. cotton-covered wire. Actually, any silk covered or enamelled wire, from 30 s.w.g. to 24 s.w.g., is suitable. If heavier wire is used then larger discs will be required.

Each disc is about 4in. in diameter, and can be stout cardboard or thin paxolin sheet. Seven slots, each about 18in. wide, are cut about 1in. deep.

Pass the wire through two small holes, and wind in and out of each slot in turn, as winding progresses. This results in half the turns lying on one face of the disc, and half on the other face, crossing over in the slots. Each coil has about 40 turns, the wire being finally anchored through two small holes, leaving the ends long enough to reach to the terminals.

With such a circuit, the parallel capacitance is mainly due to that of the aerial and earth. As a matter of interest, "tuning" coverage was tested with a signal generator, and was 1300-1700kHz with a 25pF aerial/earth system, 850-1100kHz with 100pF,

and 550-750kHz with the aerial/earth placing about 250pF across the receiver.

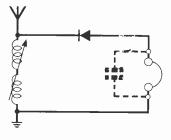
With such a receiver it was quite usual to adjust the number of turns to suit the aerial, or wavelength of local stations.

One coil is fixed on a small bracket. The other is bolted to a strip of material about lin. long, secured to a threaded rod by lock-nuts. The rod runs in two brackets, and is rotated by a large terminal head or small knob. "Tuning" is accomplished by swinging one coil over the other thus varying the effective inductance.

The base is varnished plywood or ebonite or paxolin about 6×6 in. Strips raise it about $^{3}4$ in. to clear the terminals projecting underneath.

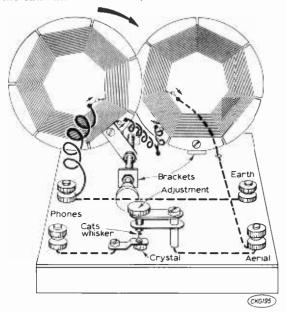
DETECTOR

A strip of brass about $1 \times^3$ ein. (from an old lamp battery) is bolted to hold the crystal firmly, a corner or point on the latter projecting up through a hole in the strip.



It could hardly be more simple! The suggested additional 'condenser' shown dotted can be about 1000pF. It will be cheaper to buy it rather than to follow the method described in the text for making a 'condenser'!

A 1½in. bolt with spacers holds a similar strip about 1¼in. long to which the catswhisker is soldered. Above this strip is a stouter strip of metal, and a screw or terminal head runs through a threaded hole in this, or through a nut soldered to it. Turning this screw or terminal adjusts pressure of the catswhisker on the crystal.



This drawing, showing the constructional details of the crystal set, should not be difficult to follow.

The catswhisker is made by winding about 1½in. of copper, tinned-copper or brass wire, about 34 s.w.g., on a small drill about ½in. diameter, and then stretching the winding slightly. These bygone detectors used wonderful combinations of crystals and even gold-tipped whiskers, all of which appeared to give about the same result. There is great room for experiment here, using various kinds of wire, or soldering the crystal (use Woods metal) or packing it with metal foil. All these trials will probably give a detector of about equal efficiency to that shown, which is about the same as a modern crystal-diode, when a sensitive spot on the crystal has been found.

AERIAL AND EARTH

The aerial ought to be at least 25ft long, and preferably over 50ft. Maximum range is usually considered to be about 25 to 150 miles from a major transmitter, anything over about 50 miles generally needing some 50-100ft or so of outside aerial, 20ft or more high.

The earth lead runs to a metal spike or other earth rod in damp soil. With a 180ft aerial ample volume was obtained by the author some 25 miles from a transmitter. No earth was used but the earth is usually desirable.

The headphones ought to be good-quality, sensitive headsets of about 500Ω to $2,000\Omega$ resistance, so as to give best volume with the rather limited output of the crystal set. Strictly speaking, a capacitor of about 1,000pF should be connected across the headphones but it was not always fitted in the early days. A description of how to make such a "telephone condenser" appeared in "Amateur Wireless" in 1922.

"The telephone condenser is made up of twenty-five small sheets of tinfoil measuring 1½in. by ¾in. with a small strip left at one corner to make a lug. It is built up by placing the strips of tinfoil, with strips of waxed paper in between, with the lugs alternately at one end and then at the other. When the condenser has been built up the ends of the tinfoil should be carefully soldered together by means of a blob of solder, two pieces of cardboard being placed either side of the condenser and a length of linen tape wound round to keep the whole together. The condenser should then be immersed in molten paraffin and allowed to set in a solid block."

The "components list" is short and sweet! One galena crystal (ref. X6), four 4BA brass terminals, 2oz. reel of 26 s.w.g. enamelled wire and a length of 4BA brass studding, all of which are obtainable from Home Radio (Mitcham) as a "kit" for 96p which includes post and packing.

TELEVISION

SEPTEMBER ISSUE

SIMPLE CROSSHATCH AND DOT GENERATOR

Constructional details for this essential item of colour TV servicing equipment. The instrument is cheap enough to be of interest to enthusiasts for do-it-yourself convergence adjustments. Notable features of the new design are: a choice of four patterns; miniature size made possible by the use of TTL MSI integrated circuits; sync amplifier for stability; suitability for use with any 625-line set with only two easy connections.

TV NOISE FIGURES

Noise factor, signal-to-noise ratio, front-end noise, aerial noise, valve and transistor noise—are you sure of yourself in this important area? If not read Gordon J. King's clear presentation of the subject this month. The usable sensitivity of a television receiver is dictated by its noise performance so this is a subject of practical importance—especially for fringe area reception. The article shows how a decision can be made on the type of aerial required and the improvement that can be expected by using an aerial preamplifier.

COLOUR RECEIVER

This month the timebase board—complete sync, field timebase and line oscillator circuits plus the line output stage with the exception of the line output transformer assembly. With board layout.

SERVICE NOTEBOOK

More items from G. R. Wilding's day-to-day experiences of TV fault conditions.

PLUS ALL THE REGULAR FEATURES

ON SALE AUGUST 21

TRANSMITTER for 2 metres

PART 1 F. G.RAYER G30GR

THERE is quite a lot of interest in 144MHz equipment, especially among the Amateur Class B licencees who are permitted to transmit on 144MHz and above. Apart from this, some of the advantages of the use of this band may be noted. There is considerable frequency space, the 2-metre band being 2MHz wide, from 144 to 146MHz. A large space for an aerial is not necessary, aerials for 144MHz usually being single-mast affairs resembling those used for TV purposes. Many stations use low power, in the 10 to 15 watt region, and crystal control of the transmitter is quite usual, which makes the equipment relatively simple to build and operate. There is also a very good chance of clear, solid QSO's, free of all the interference that makes 160m or 80m so uncomfortable.

The transmitter described here runs about 10 watts input, so it can be run from a power supply and modulator such as may already be in use on a 160m transmitter. If the inductors are made and adjusted as described, it will be found virtually impossible to make the transmitter operate outside the permitted band.

CIRCUIT

Fig. 1 shows the complete transmitter circuit which uses four valves. Operation of the various stages will probably be more clearly understood if they are dealt with separately.

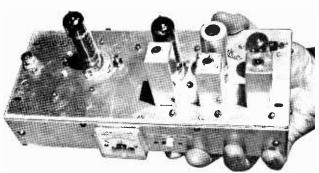
V1 Oscillator. This is a simple crystal oscillator using a 6C4 in which L1 is tuned to the same frequency as the crystal, approximately 8MHz, and V1 cannot be made to work on any other frequency. The test point TP1 allows the grid current of V2, produced by grid rectification of r.f. from V1, to be measured, to facilitate the initial tuning of L1.

V2 Tripler. The 6BH6 receives input at 8MHz, and triples this to 24MHz, to which L2 is tuned. Point TP2 allows a meter to indicate grid current of the following stage, as an aid to the initial tuning of L2. The wrong multiple, if selected by L2, would be 16MHz or 32MHz, and these frequencies, which would result in incorrect operation cannot be tuned by L2.

V3 Tripler-Driver. This is a double beam tetrode QQVO3-10. The first section receives drive at 24MHz and triples to 72MHz, to which frequency L3 and L4 are tuned. Grid current of the following section of this valve can be monitored at TP3. L3/L4 cannot be tuned to a wrong harmonic.

The second half of V3 receives drive at 72MHz from L4, and doubles to 144MHz, to which L5 is tuned. Trimmer TC1 aids in balancing the centre-tapped circuit L5.

V4 Power Amplifier. L6 is tuned to 144MHz and grid current of the QQVO3-10 develops about 30-60V bias across R14. The switch S1a/b allows grid current



to be monitored, which is essential. For grid current indications, the 5mA meter reads directly.

Cathode current for V4 flows through R19. With the meter switch in the second position, R20 is in series, and the meter reads 0-0·5V, which corresponds to a full-scale deflection of 50mA, so that the cathode current of V4 can be measured.

A separate h.t. connection is provided for V4, so that modulation can be applied for speech transmission. This type of valve does not need neutralising although acting as a straight-through 144MHz amplifier. L7 is tuned to 144MHz, and output is from the link coupling L8, TC2 allowing adjustment of loading. The transmitter will work directly into a co-axial feeder to a dipole or similar aerial.

With a 280V supply for the early stages, it was found quite easy to obtain more grid drive than was required. Running the p.a. with about 10 watts d.c. input will light a 6 watt lamp used as a dummy load very well, so a good level of efficiency is obtained.

INDUCTORS

L1 and L2 are wound on $1^{7}_{8} \times {}^{1}_{4}$ in. dia. formers and L3 and L4 on a single $2^{3}_{8} \times {}^{1}_{4}$ dia. former. The formers fit in ${}^{3}_{4}$ in. square screening cans and are mounted by 6BA bolts which pass through lugs on the cans and into tapped holes provided on the square bases of the formers. Fig. 2 shows winding details. To make L1, bare and solder 32 s.w.g. enamelled wire in the tag used for h.t. Leave ${}^{1}_{2}$ in. clear space, then wind on 50 turns close wound. Secure the ends with polystyrene cement or similar adhesive, clean the wire end and solder it in the A (anode) tag. The wire ends should slope down clear of the windings, and clear of the screening can.

If surplus or other formers or cans are used, these need to be long enough to avoid having the windings near the chassis or top of the can. It might also be necessary to make sure that the cores of surplus coils are suitable for the frequencies involved.

L2 is wound in the same way, and has 15 turns of 24 s.w.g. enamelled wire, close wound.

L3 and L4 each have 5⁵4 turns of 24 s.w.g. enamelled wire, close wound. A gap of ¹gin. is left between the two coils. Begin winding at G, Fig. 2, and finish this coil at the bias circuit lead, R11. Then begin L3 at the h.t. positive end, continuing to wind in the same direction as for L4, and terminate this at end A (anode). The long former listed has soldering eyelets, and the wire ends pass down through these and soldered but not cut off. The leads emerge as in Fig. 5. They are cleaned and left long enough to reach the required points. To avoid possible short circuits pieces of sleeving are put on the pins of L1 and L2, and the wire ends of L3 and L4. For easy identification, the wire ends of the latter can

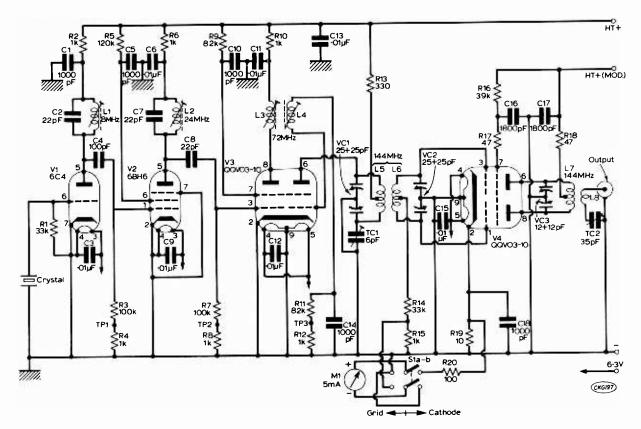


Fig. 1: Circuit of the 2-metre transmitter which uses 8MHz range crystals in the oscillator stage.

have coloured sleeving, as follows: Black—grid; Green—bias resistor; Red—h.t.; Yellow—anode.

Place each former in its can with the threaded holes matching the lugs, which are bent over. To prevent the cores moving, a piece of a thin elastic band can be put into each former, before screwing in the core.

The windings for L3 and L4 may be tuned to resonance with the cores only using stray circuit capacitances. If cores are employed, the core material must be good for 72MHz.

The 144MHz inductors are shown in Fig. 3. L5 and L6 are wound with 18 s.w.g. wire, and L7 with 16 s.w.g. wire. Sleeving should be put on the wire for L5 before winding to prevent possible contact with L6.

L5 has five turns wound on a suitable object, such as a pencil, so that its outside diameter is $0 \cdot 4in$. and turns are spaced to occupy $0 \cdot 6in$.

L6 has six turns, 0·3in. outside diameter, and is 0·7in. long. L7 is four turns, spaced to occupy 0·5in., and its outside diameter is 0·6in. L5 has a centre tap soldered on the top of the central turn, while L6 and L7 have taps on the bottom of the centre turn. The loop L8 is a single turn of 16 s.w.g. or similar wire, placed in sleeving, and with ends shaped to reach the coaxial socket and loading capacitor.

CHASSIS ASSEMBLY

The chassis and screen across V4 valveholder consists of six "universal chassis" flanged members. The top, 10×4 in., is prepared for the valveholders etc. as in Fig. 4. The screened coils need four holes to clear the pins. two holes for 6BA bolts, and a central hole under L3/L4 to reach the bottom core.

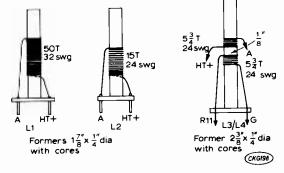


Fig. 2: Winding details of the coils for the oscillator and multiplier stages.

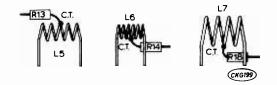


Fig. 3: Construction of the coils for the power amplifier. Put sleeving on L5 before winding.

A $4 \times 2in$, member is bolted **inside** each end flange of the top. The third $4 \times 2in$, member is cut so that it can be placed across the holder for V4 as in Fig. 5, clearing tags 4 and 5 (heater). It is bolted to the top at the valveholder.

The 10×2in. front member has the top flange cut off, and takes the meter and 2-way switch in the positions shown in Fig. 5. It is bolted to the chassis

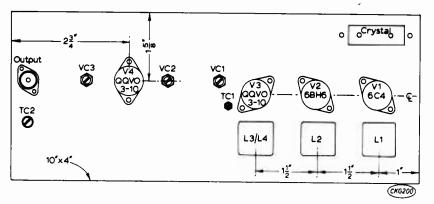


Fig. 4: Marking-out details on top of the chassis. Holes for the single tag stand-offs can be drilled as construction proceeds.

top, both ends, and the screen. The back is fitted in the same way.

It is best to prepare the front and back but leave them off until most of the wiring is completed.

WIRING

Fig. 5 shows wiring and other details. All by-pass capacitors, and particularly C10 to C19, should be connected from the various tags to adjacent chassis tags with the shortest possible leads. Resistors R13, R14 and R18 are soldered with very short leads from resistors to centre-taps, and R17 is wired directly to pin 7 of V4.

Capacitors VC1, VC2, VC3 and TC2 may be fixed by the spindle bushes, or by bolts. In either case the rotors are grounded by stout leads to near-by chassis tags. All r.f. leads in the 144MHz circuits must be stout and no longer than necessary.

Single tag stand-offs were used to provide anchor points for R2, R6, R10, etc., and also for the grid current test points TP1, TP2 and TP3.

A 4-way tag strip is bolted as shown, for flexible supply leads or a 4-way cord. This provides 6.3V, chassis or common return, h.t. positive, and modulated h.t. connections to the power supply and modulator.

CIRCUIT ADJUSTMENTS

When an indication of grid current has been obtained on the meter this can be used for all tuning adjustments. However, it is not very likely that this indication will be obtained immediately, so it is generally necessary to adjust the stages one at a time, as described below.

A 6.3V 2A supply is necessary for the heaters. A 250-300V supply at about 60mA is required for the h.t. for V1, V2 and V3, actual current depending on the voltage and adjustments. Temporarily leave h.t. to V4 (Mod.) disconnected.

Crystal. Multiplication is $3\times3\times2$, or 18 times. Crystals in the region of 8MHz are thus necessary. To

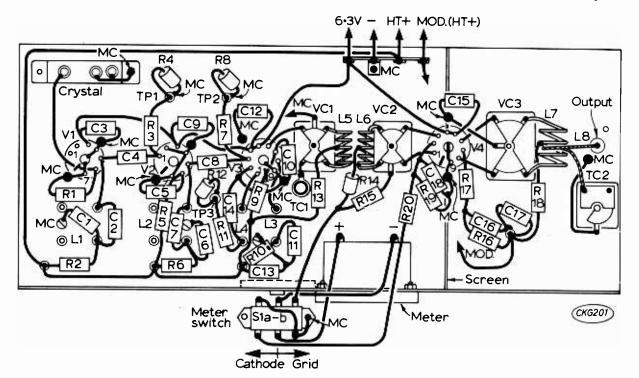


Fig. 5: Wiring underneath the chassis. Stand-off insulators are used for the test points TP1, 2, 3 and to support resistors R2, 6 and 10.

* components list

Resis	tore					
		DO	1kΩ } W	R15	1kΩ	11//
R1	33kΩ ½W	R8				-
R2	1kΩ ⅓W	R9	82kΩ 1 ₩	R16	39kΩ	
R3	100kΩ ¥W	R10	1kΩ ⅓W	R17	47Ω	₹W
R4	1kΩ ¥W	R11	82kΩ ½W	R18	47Ω	₹W
R5	120kΩ ¥W	R12	1kΩ ǯW	R19	10Ω	₹W
R6		R13	330Ω ½W	R20	100Ω	₹W
D7	100kg 1W	R14	33k0 ÎW			

Capacitors

apa	CILUIS			
Č1	1000pF	350V disc	C10 1000pF 350V disc	
C2	22pF	SM	C11 0.01µF 350V disc	
C3	0.01µF	12V disc	C12 0·01µF 12V disc	
C4	100pF	SM	C13 0.01µF 350V disc	
C5	1000pF	350V disc	C14 1000pF 350V disc	
C6	0.01µF	350V disc	C15 0.01µF 12V disc	
C7	22pF		C16 1800pF 1kV disc	
	22pF		C17 1800pF 1kV disc	
C9		12V disc	C18 2000pF 350V disc	

VC1-VC2 25+25pF butterfly capacitor (Jackson type C713 (0.015" gap)

VC3 12+12pF butterfly capacitor (Jackson type C713 (0.045" gap)

TC1 6pF tubular trimmer (Home Radio VC88B)
TC2 35pF pre-set air spaced

Valves

V1	6C4	V3	QQVO3-10
V2	6BH6	V4	QQV03-10

Chassis

"Universal chassis" flanged members.
10 x 4in. type CU58A (1), 10 x 2in. type CU139 (2),
4 x 2in. type CU133 (3) (Home Radio)

Miscellaneous

M1, 5mA miniature panel meter. Valveholders B7G, skirted with screening cans (2) B9A (2). S1a-b, 2 pole 2 way slide switch. Crystal 8MHz (see text) and holder. Coil formers type CR14 (2), CR16 (1), screening cans CR15 (2), CR17 (1). Cores type CR19 (4) (Home Radio). Tag strip, stand-off insulators etc.

find the output frequency with a given crystal, multiply the crystal frequency by 18. Alternatively, to obtain a crystal to operate on a particular frequency, divide the wanted frequency by 18.

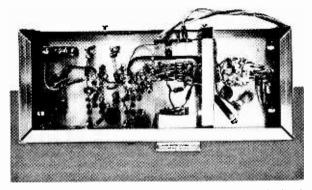
Oscillator. A suitable tool to adjust the coil cores can be made by filing flats on the end of a plastic knitting needle. Clip a multi-range meter or other suitable instrument from tag TP1 to chassis. A 5mA or similar range is suitable. Or if a voltage range is used, 1V will indicate about 1mA, and so on.

Rotate the core of L1 for nearly maximum grid current at TP1. Do not tune L1 exactly for peak current, or it will be found that the crystal oscillator will not start, when switched on. This is usual with this type of oscillator.

Tripler V2. Transfer the meter to TP2 and adjust the core of L2 for best grid current.

Tripler-Driver V3. Transfer the negative meter lead to TP3 and rotate the cores of L3 and L4 for best grid current. These are quite critical, and interact to some extent.

Power Amplifier. Sla/Slb should be set to read grid current. Slowly rotate VC1 and VC2 for best grid current. When a reading is obtained, the test-



View underneath the chassis. The vertical screen isolates the input and output circuits of the p.a.

meter can be unclipped, and cores L1 to L4 can be touched up, as necessary.

Should a peak in grid current arise with VC2 fully open, slightly stretch L6, and re-adjust VC2. On the other hand, if VC2 is fully closed, press the turns of L6 a little closer together. This also applies to L5 and L7.

Unscrew TC1, and peak grid current with VC1. Slowly screw down TC1 meanwhile re-adjusting VC1 for best grid current. This will give a peak somewhere the middle setting of TC1, but is not critical.

Final grid current can be adjusted by varying the spacing between L5 and L6, and anything around about 1.5mA is suitable.

It is best to make the first output test into a lamp load. This can be a 12V 6W bulb, connected to a co-axial plug to connect to the output socket.

Open TC2 to minimum capacity and switch the meter to read cathode current. Provide about 300V or so at the "Mod" connection. Rotate VC3 for a dip in cathode current, which should cause the bulb to light. The meter can then be returned to "Grid" and VC2 readjusted (also possibly VC1).

LOADING

The loading of V4, and thus the final current drawn when VC3 is tuned to resonance, is adjusted by rotating TC2, and if necessary moving L8 in or out of the coil L7.

Maximum listed ratings for the QQVO3-10 are 300V for anode and 175V for the screen grid, with an anode current of 76mA (d.c. input 22W), 3mA grid current, resulting in an output of about 14W. The power supply and modulator used was more suitable for about 40mA at 280-290V or an input of about 12W. Screen-grid current is about 3mA, and the meter fitted shows the total cathode current, anode current alone being a little less. If the meter is to read 0-100mA, for larger inputs, R20 should be 200Ω .

The modulator should be able to furnish at least one half of the input to V4 in watts—say 5W for 10W input to V4. A 12AX7 followed by a single 6BW6, as a Class-A modulator (as used for a 160m transmitter) was found to work well with an input of up to about 10 watts to V4.

Next month we shall describe a power supply and modulator for use with this transmitter. It can also be used as a conventional audio amplifier or as a power supply for a higher powered transmitter.

ARKING LIGHT

J. N. JONES

A LTHOUGH the regulations regarding the lighting of cars after dark have recently been relaxed, there are many areas and situations where parking lights are still required. If your car has to display lights regularly, it is a chore having to go out to turn them on—and there is always the risk of forgetting to do this.

For quite a small cost (only about £1·25), it is a simple matter to arrange for a parking light to come on automatically when the light level falls below a certain level and to switch itself off at daybreak. This obviously conserves the battery, especially during the winter (with the long nights), the time when the battery needs to be in peak condition. Such a device is especially useful when the car is used only occasionally.

+12V €12V R2 100k 50k Tr1 R4 2N2906 1.8M 2N3704 11 6 741 5 6 1 LDR ORP12 ≤R3 100k 12V 3W Lamp

Fig. 1: The circuit for both positive and negative chassis cars. Where there is a variation, +ve details are in a squared box, -ve details in an oblong box.

changes over this range, the transistor is switched from completely off to fully on, causing the bulb in the collector to light.

The purpose of the positive feedback action of R5 is to prevent possible oscillation and to ensure that the switching action is fast enough to prevent destruction of the transistor which can only handle the current in fully on or fully off states.

The negative earth circuit functions in exactly the same way but the transistor type needs to be changed and the voltages reversed.

THE CIRCUIT

The circuit shown in Fig. 1 is suitable for cars with either a negative or positive chassis with appropriate modifications. The circuit makes use of an LDR (light dependent resistor) and the type 741 operational amplifier IC. The latter is an extremely useful and well protected differential amplifier which is available for as little as 34p—low enough to promote serious consideration in simple circuits such as this. The main features of the circuit are the low cost, already mentioned, and the low current drain in the standby position; in daylight the current drain is only about 2.5mA.

The positive earth circuit functions as follows: R3 and R4 fix the non-inverting input of the IC to a potential of about -6V. This is slightly modified by the positive feedback action of R5, but only by a fraction of a volt. VR1 is set so that at lighting-up time the potential of the inverting input, going negative with falling light levels, is just passing through -6V. This causes the inverting input to become negative with respect to the non-inverting input, which in turn causes the output potential to change from about -10V to about -2V. The values of R6 and R7 are selected so that when the output

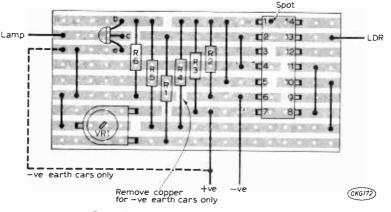


Fig. 2: Component layout on Veroboard. Note that for negative chassis cars R2 and R3 reverse notation but, as they are of the same value, no circuit changes are necessary.

CONSTRUCTION

A suggested layout on Veroboard is given in Fig. 2, with the changes from one chassis potential to another given. This circuit can be mounted anywhere in the car but the LDR has obviously got to face through the window. It should not face the parking light; if it does the unit will "oscillate" as the circuit cannot differentiate between the light it is meant to switch on and daylight!

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heat sink, solder and booklet"How to Solder



PRICE £2.40

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185	-22		-76		-10		+38		-82		-80
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384	-26	30F5	-64		-54		-50		-84		-50
3V4	-47	30FL1	-65		•30		-36		-38		·45
5U4G	-81	30FL12	-69		-22		-29		-38		-82
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5Y3GT	-34	30L1	-29	EBF83	-39		-43	PCL800	.75		-85
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6/30L2	· 54	30L17	-67		-17	EZ80	-22		-77		-58
6AL5	•11	30P4	-65		-20	EZ81	.23	PEN36C	-70	UCH81	-32
6AM6	-13	30P12	-69		-35	EZ90	-23	PFL200	-52		-32
6AQ5	-22	30P19	-65		-84	GZ30	-34	PL36	-49		-55
6AT6	.20	30 PL1	-60				-40	PL81	-44	UF41	-52
6AU6	-20	30PL13	-89	ECF80	-31	GZ34	-48	PL81A	-47	UF89	-30
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6BW7	.52	35Z4GT	-25	ECH81	-29		-63	PL500	-63	UY41	-89
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6F25 6J7G	-53	B349	-65	ECL80	-35		-87	PX 25	-95	W77	-43
6K7G	-24	B729	-62	ECL82	-31	P61	•40	PY32	.55	Z77	-22
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6Q7G	·17	CY31	+30	EF39	-38	PC86	-47	PY81	-25	AC107	-17
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12AU7	-20	DL35	-40	EF184	-31	PCC805	-98	U26	-56	OC44	.12
12AX7	-22	DL92	-28	EH90	-34	PCF80	.28	U49	-84 -58	OC45	-12
19BG6G		DL94	-47	EL33		PCF82		U52	-31	OC71 OC72	-12
20F2		DL96	-38	EL34	-45	PCF86	-48	U78	-24	OC75	-12 -12
20P3		DY86	-24	EL41		PCF800			-59	OC81	-12
				EL84		PCF801		U193	-42		-12
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A good turn

With reference to the insertion in your CQ? Column June '72 you will be very pleased to know I have received a copy of the book I was looking for, "Simple Radio Circuits."

The reader who sent me this is in search of some detail concerning "Modification details of any ex-military receiver" if you could help him in this matter through your column I shall be very grateful.

The name of the writer whose request I mention is Alan Salisbury, 28 Dyke Street, Brymbo, Wrexham, N. Wales LL11-5 AH.—

J. Wheelton (Staffs).

Greek notes

With reference to W. D. Logan's letter in the May editorial; I wouldn't try to disappoint him, since I am too a vintage radio enthusiast and collector but I think that the early days of Radiotelegraphy were not so romantic and sentimental as he often dreams.

Until high frequency communications were developed, signals were pushed long distances by brute force. To generate the necessary power, great sparks buzzed, molten arcs sizzled, and mechanical alternators revved up madly to as high as 20,000 r.p.m. their rotators yearning to fly apart.

To make these rock-crushers work, an early operator had to be electrician, steeple jack, rigger, mechanic and oiler.

To turn on some arc transmitters he also had to be something of an acrobat. Trouble was when the arc chamber was filled with hydrocarbon gas—if there also was the right amount of air in the chamber, the gas blew up and the carbon was fired back to the operator.

Because receivers had little or no amplification, the operator had to develop sensitive hearing. According to the expression, a man had to crawl into the headphones in order to copy weak signals!

Stations ordinarily were hundreds of miles apart and sending had to be slow. A distant spark

signal, for example, sounded like a gnat sighing through a screen and the burbly signal of a far away transmitter might be weaker than the noise, the methodically taped dihs and dahs audible only through holes in the static.

So all important was an individual's adeptness that every operator struggled to avoid ending up with the trade's own brand of punch-drunkenness, a tin ear and a glass arm.—Chris Petsikopoulos (44 Atlantos str. P. Faliron, Athens, Greece).

CO! G. Saunders

While reading through the C.Q. section of June Practical Wireless I came across a book wanted by a Mr. George W. Saunders with no address given. Could you please ask him to write to me if he is still looking for this book. The book I have is "Mullard", Circuits for Audio Amplifiers; reprinted June, 1966. Please ask him to write to the address below for details.—E8097840 SAC Lim Hangar 91, RAF Brize Norton, Oxford, OX8 3LX.

Cassette decks

With reference to the letters of V. S. Watts and A. R. Knight, I have a source of Philips type cassette mechanisms in about three different models. They are expensive (nominally £25 each) and if Mr. Watts and Mr. Knight or any others contact me I will endeavour to help them.—S. R. Beeching (Consortium Engineer) (Bishop Grosseteste College Lincoln).

Egg-shapped

I have just recently invested in a Henry's Radio catalogue, and I note that on page 274 it says at the bottom, "Note, deliveries of speakers, particularly elliptical tapes, can sometimes be difficult." Is this a new type of tape? and are these tapes supplied on elliptical reels?

I should be interested in your comments.—F. G. Jennings (Sussex).

[This appears to be a type error! —Editor]

Bang!!

Although it is too late to enter for the competition for the most interesting find made by the PW Treasure Tracer Mk I, I thought you might be interested to know that I found a bomb in my garden recently. The police think it is some kind of mortar bomb which has been fired but has not exploded. When I think how roughly I handled it compared with the way that the three policemen handled it I nearly died of fright.

The bomb is made by ICI and is about 9 inches long and has corroded terribly. It was buried about 2 inches in the ground and just to think that I walked over it many times a day. Apart from the mortar bomb and tail fin I also found a spent cartridge about three feet away from it.

If I had dropped this long cylindrical bomb I dread to think what would have have happened to me and my neighbour who was watching.—N. Moyes, (Croydon, Surrey).

Equipment

Mrs. V. E. Whetstone has informed us that she has a large amount of radio and television equipment that belonged to her late husband. Items include Radio and Television Servicing volumes 1-8, numerous issues of Practical Wireless and Television dating back to the 1940s, loudspeakers, oscilloscopes, valves, tubes, transformers etc. and many items of test equipment including meters. There are many line output and e.h.t. transformers, resistors, capacitors, coils, valveholders and television i.f. panels. There is a sound measuring device together with time switches, turntable and many other items of useful equipment, including CME 2301 and CRM 172 television tubes to name

If any readers are interested in making offers for any of these items (and Mrs. Whetstone has expressed a wish that she would like radio clubs to have some of the equipment) would they please phone 01-500-1513.

Project Autumn's presentation

T a presentation luncheon in London recently John Thornton Lawrence was presented with the PW Designers Trophy for 1971 by Editor Norman Stevens. The winning article "A Digital Frequency Counter/Timer" was published in PW from September to December 1971 and was considered by a panel of judges to be the best entry in the PW "Project Autumn" competition.

John's interest in radio began at school. An apprenticeship in radio engineering at an electronics factory on the inspection side gave him a good insight into multiband short wave receivers. In 1947 he switched to the radio trade as the service manager for a company specialising in radio, sound systems, disc recording and TV. At home, he designed and built a 12in. TV set together with an oscilloscope and pulse generator to align the set before regular BBC

transmissions started.

In 1953 he became GW3JGA operating mainly on Top Band and 2 metres, but in 1958 as GW3JGA/T made the first GW two way television contact, with GW3FDZ/T, at a distance of 18 miles. He was a founder member of the Flintshire Radio Society and was the RSGB's Region 11 representative for a period. In 1960 he joined the Electronic Engineering Department of the University College of North Wales where he now holds the post of Senior Scientific Officer, and incidentally, founded the UCNW Amateur Radio Society whose station GW3UCB is frequently heard in contests on all bands.

John's interest in amateur TV expanded and in 1962 he completed a sequential colour system and transmitted what are believed to be the first amateur colour TV pictures in this country. With later transistorised equipment he achieved a two way TV contact with the Isle of Man at a range of 79 miles. Among other qualifications John is a Fellow of the Society of Electronic and Radio Technicians and a member of the Royal Television Society.

The pay-off! JTL (left) gets his award from WNS.





Practical Wireless Designer's Trophy 1972

To encourage new authors, entries for the 1972 Trophy will be restricted to readers who have not previously had an article published in PW. This leaves the field wide open for those wanting to try their hand at writing technical constructional articles. Contestants will not be in competition with well-known authors, only with other newcomers, so the cup can only be won by a new writer. It Could Be You.

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- The winning entry will be chosen by a panel of judges from among articles published in issues of PW dated September 1972 to August 1973 inclusive. The Editor's decision on all matters arising will be final.
- The winner of the competition will receive and retain outright the PW Designer's Trophy 1972. Other prizes will be awarded to the best runners-up. Articles will be paid for shortly after publication.
- The competition is open only to authors who have not previously had any work published in PW.
- 4. Articles submitted for the competition should conform to the general style of material published in PW and must describe the operation and construction of a piece of radio, audio or test equipment that has been designed and built by the author.
- 5. Articles should, preferably, be typed using double spacing, leaving wide margins, on one side only of each sheet. Circuit diagrams and any other drawings must be separate and numbered to agree with the text. Author's roughs must be clear enough to permit re-drawing. Components list must also be separate and laid out to the standard PW format.
- Photographs of the equipment are desirable and should be in black and white, sharp and clear. Photographs may be identified by sticking a label on the reverse instead of writing on the back of the photograph itself.
- Components used in the design must be readily available from retail sources.
- 8. Articles should be sent to the Editor, Practical Wireless, Old Fleetway House, Farringdon Street, London, E.C.4. Authors will be advised as soon as possible of the acceptance or rejection of their articles. Equipment, the subject of an article, must not be sent to the Editor until advised to do so.
- Employees and staff of PW are not eligible for entry to this competition.

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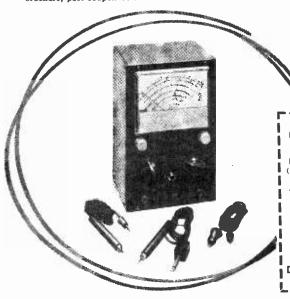
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These give control on N.S. & E.W. directions, movement of control stick operates I pole c/o micro svt in all four directions, also 2 more micro swts are operated when stick moved from neutral position. Ex equipment. Price £1:00 plus 20p post.

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3 types available as follows: 20pf max air spaced ceramic base, screw-driver adjustment. Price 10 for 50p or £4 per 100, 50pf max air spaced ceramic base, 5 for 50p, 10 to 110 pf ceramic dielectric 5 for 50p. All new, unused and post paid.

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Part of Aircraft Nav display center reading type scale marked L & R, M.C. meter, as two coils wound on magnet, this enables reading direction and amount to be controlled by two separate circs. Supplied with plug conn. in new condition.

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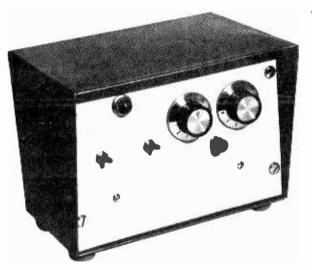
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For example tested L.O.P.T. £1-45. Fireball Tuners with Valves 75p. Droppers

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this set; Reclaimed T.V. Tubes. All with 12 months guarantee. AW43/88, £1-50. AW43/80 £1-50. MW43/69 £1. Special Offer, Brand New Brimar Tubes C17PM £1. Many older types in stock. Carriage and ins. on any tube £1-50. Valve list ex equipment. All valves tested on a Mullard valve tester before despatched. 3 months guarantee on all valves. Single valves P/P 3p. Over

ARP12 EB91 EF80 EF85 EBF80 EBF89 ECC81 ECC82 ECC83	5p 4p 8p 12p 12p 12p 10p 12p	PCC84 PCF80 PCL82 PCL83 PCL84 PL36 PL81 PY81	5p 5p 12p 12p 12p 12p 17p 8p	U191 U251 6BW7 6U4 20P1 20P3 20D1 30P4	20p 12p 10p 10p 20p 10p 10p
ECL80	8p	PY33	17p	30F5	10p
EF91	4p	PY82	8 p	30P12	20p
EY86	20 p	PL82	8 p	30FL1	20p
EP50	60p per doz.	PL83	₽ p	6/30L2	20p



Precision ENLARGER TIMER

R.A.BOTTOMLEY

ost electronic enlarger times utilize the principle that a capacitor, when connected to a supply voltage by way of a series resistor, takes a certain time to reach a predetermined voltage. Some form of voltage level detector is then used to indicate when this level has been reached and then the capacitor is discharged in readiness for a further timing cycle. The voltage level detector can take various forms but nowadays most enlarger timers are designed around the unijunction transistor. These timers are reliable but they suffer from one or two disadvantages.

Due to the fact that the UJT requires a minimum amount of emitter current to trigger it, a limitation is imposed on the amount of resistance that can be used in the timing circuit. In order, then, to get a reasonably long time interval a large value capacitor has to be used and this is invariably of the electrolytic type which is not the most stable of components. The variable element is usually a potentiometer and this requires a hand-calibrated scale which is a drawback to those of us who lack that particular drawing skill! Even should we possess this skill there remains the difficulty of reading the scale in the subdued light of the darkroom. It was this last factor, more than any, which was mainly responsible for the present design.

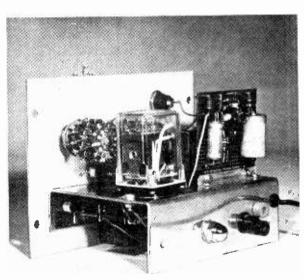
Any time interval, within the range of the instrument, can be set on two decade switches by feel; a simple matter, akin to stopping down a lens by the feel of the click stops. Indeed the operation of the instrument has proved to be so simple that no panel marking has been considered necessary.

DESIGN CONSIDERATIONS

The use of decade switches imposes certain restrictions upon the design of the instrument. If a high degree of accuracy is to be achieved, there must be no leakage in the timing capacitor. This rules out the use of a capacitor of the electrolytic type, although a tantalum type should be better in this respect. Since a fairly large value capacitor will still be required, a low voltage polycarbonate type seems to be indicated and in the prototype a 6.8 pc component is used. This particular value was largely dictated by size and cost.

In order to obtain fairly long time intervals even

with a capacitance value of this order, the resistive component of the time constant amounts to several megohms. This, as previously stated, rules out the use of a UJT and in the present design a Schmitt trigger is employed as the voltage level detector. Since this, too, would present a low resistance load to the timing circuit an f.e.t., connected as a source follower, is interprosed between the timing circuit and the Schmitt trigger.



A rear view of the completed prototype.

CIRCUIT DESCRIPTION

The circuit diagram, Fig. 1, shows the state of the push-button switch and the relay contacts immediately before the timing cycle. The timing capacitor, C1, is fully discharged by relay contacts RLA/1 and the gate of Tr1 is at zero potential. Only a small voltage appears at the source due to the self-biasing effect of the small amount of drain current which flows through the resistor chain R22, VR1, R23. As a

Fig. 1: Above is the complete circuit of the timer with the power supply circuit shown below.

consequence, Tr2 is off, Tr3 is on and Tr4, which is a PNP transistor, would also be on were it not for the fact that the relay contacts RLA/2 and the pushbutton switch S3 are both open. When the pushbutton switch is actuated, Tr4 conducts, the relay coil is energized, relay contacts RLA/2 close and the relay is held on by these contacts.

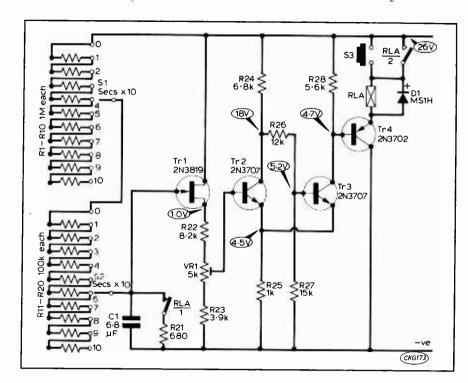
Simultaneously. relay contacts RLA/1 open and allow the timing capacitor to start charging up through the series of resistors selected by the decade switches. The voltage on the gate of Trl then rises in an exponential manner and the voltage on the source follows in sympathy. When this voltage reaches a certain value, Tr2 goes into conduction and Tr3 switches off smartly due to the regenerative action which is a characteristic of the Schmitt trigger. Tr4 is also switched off and the relay falls out thus completing the timing cycle. During this timing cycle, relay contacts RLA/3 close, thus applying the mains supply voltage to the enlarger lamp.

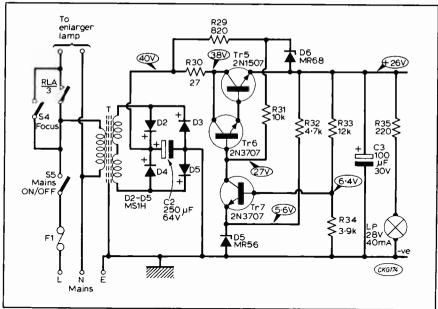
In order that the timer should give consistent accuracy it is necessary that the supply voltage be held constant. Transistors Tr5, Tr6, Tr7 and their

associated components form a fairly conventional stabilised power supply and the instrument is proof against mains voltage fluctuations over the range of 220 to 260V. Although Tr5 (2N1507) is operating within its rating, a small TO5 heat-sink is mounted on it for added security.

CONSTRUCTION

There is nothing critical about the layout and the constuctor is left to use his own ideas on this subject. Several prototypes have been constructed utilising various layouts and no spurious results have been encountered. The photographs show one neat layout which has become the writer's final choice. The





★ Specification

RANGE
0-110 seconds, in 1 second steps, selected by two decade switches.

ACCURACY
5 per cent of set time or better (see text).

"SET-ABILITY" By feel.

STABILITY Unaffected by mains voltage variations between 220 and 260V.

REPEATABILITY
Of a very high order as there is no possibility of disturbing a moving scale.

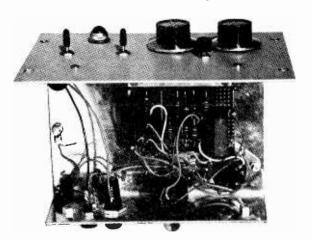
circuit has been conveniently split into two portions and each part has been constructed on a small piece of Veroboard. The one above the chassis, along with the miniature mains transformer, comprises the regulated power supply. The one below the chassis accommodates all the components, with the exception of the relay, which go to make up the timer unit itself. These boards are fixed to the chassis by means of small angle brackets fabricated from short strips of light-guage aluminium.

The decade switches, push-button switch, focus switch (which by-passes the relay contacts RLA/3 during the composing and focussing procedure), mains on/off switch and pilot lamp are mounted on the front panel. The whole is housed in a Contil 755 instrument case which comes complete with non-slip plastic feet. As its type number suggests, this case has dimensions of 7in x 5in x 5in and completes a very neat instrument which occupies little space. The decade switches, as shown on the circuit diagram, are single-pole 11-position types but in the final prototype 2-pole 2-wafer switches are used and the decade resistors are conveniently mounted between the two wafers resulting in a neat assembly.

The relay is mounted by means of its 11-pin valveholder type base. Some comment on this relay is called for. An incandescent lamp is a very non-

* components list

Resistors R1 -R10 R11-R20 R21 R22 R23 R24 R25 R26 R27 All resis	$\begin{array}{l} 100 k\Omega \\ 680 \Omega \\ 8 \cdot 2 k\Omega \\ 3 \cdot 9 k\Omega \\ 6 \cdot 8 k\Omega \\ 1 k\Omega \\ 12 k\Omega \\ 15 k\Omega \end{array}$	R28 R29 R30 R31 R32 R33 R34 R35	$5\cdot 6 \mathrm{k}\Omega$ 820Ω 27Ω $10 \mathrm{k}\Omega$ $4\cdot 7 \mathrm{k}\Omega$ $12 \mathrm{k}\Omega$ $3\cdot 9 \mathrm{k}\Omega$ 220Ω
Capacitor C1 C2 C3	6·8μF ±10 250μF 64V 100μF 30V	% 63V polycarb electrolytic electrolytic	onate
Semicond Tr1 2N3 Tr2 2N3 Tr3 2N3 Tr4 2N3 Tr5 2N1 Tr6 2N3 Tr7 2N3	8819 D1 8707 D2 8707 D3 8702 D4 507 D5 8707 D6	MS1H MS1H MS1H MS1H MS1H 6·8V 5mA Ze 5·6V 5mA Ze	
T S1, S2 S3 S4, S5 F1 LP	Relay, Radio Type MK3P Radiospares 12V type (The two 12' in series) 1-pole, 11-wa 1-pole push- 1-pole mains 1A fuse-link 28V 40mA la holder with	e type NK2. Gra Is which match s nk	transformer, re connected ee text) in pilot lamp- duations 0-10



A top view of the prototype enlarger timer.

linear resistor and it exhibits a relatively low resistance when it is cold. Thus, when it is switched on, there is a comparatively high "in-rush" current and the contacts of the relay should be such as to cope with this heavy load. This is the reason for specifying the particular Radiospares (now R.S. Components Ltd.) heavy duty relay. The mains transformer is also Radiospares type. Before leaving the subject of the relay, however, it might be as well to remark that its rating is quoted for some hundred million operations. This would amount to an awful lot of enlargements! A lighter duty relay could be, and has been, used successfully but it is not considered to be good engineering practice.

SETTING UP

The accuracy of the instrument will depend upon two factors; the tolerance of the decade resistors and the care with which VR1 is adjusted. The 5 per cent tolerance resistors should be good enough for photographic purposes but if one's pocket is deep enough or if the timer is to be used for a more exacting purpose, 2 per cent or even 1 per cent resistors could be used to improve the accuracy. The writer selected the resistors from a batch of 5 per cent carbon film resistors, using a highly accurate bridge for matching purposes.

Once the construction has been completed and the wiring checked, the instrument should be switched on. VR1 should be set with its slider at the end connected to R22 (8· $2k\Omega$). If it is set to its other extremity the voltage at this point may never rise high enough to switch the Schmitt trigger. With the decade switches set for 1 second (x10 at 0 and x1 at 1) the push-button should be actuated and, if all is well, the relay will close and fall out again after approximately one second

mately one second.

The decade switches should then be set for a period of 30 seconds (x10 at 3 and x1 at 0), the push-button should be actuated and the timing period should be measured and found to be something less than 30 seconds. If this is so, indicating correct operation, the slider of VR1 should then be moved away gradually from its end stop until a position is found where the contact closure is as near to 30 seconds as possible. Once this position has been found, the potentiometer should be left undisturbed and all the other timing periods as set on the decade switches should fall into place.



OST readers will have seen and tried one of those games where you have to pass a small metal loop over a bent wire, trying to avoid touching it. When the two touch, a bell rings.

Our Buzz-Bar is similar in many ways but it has extra facilities which make it much more fun to use. The level of skill, unlike the simple version, can be varied so that children stand an equal chance with their elders. A lot of fun can be had from a project of this type and it makes an attractive sideshow at a fete or bazaar—especially as it is a bit out of the ordinary.

The main difference between ours and the more conventional set-up is that the "wand" (the small metal loop) can touch the bar for a limited period or number of times. This allows the shape of the wire to be bent into really weird and wonderful shapes and it can be made so that a mistake-free run is virtually impossible.

We have said a limited number of times. Each time contact is made, the electronics part of the circuit adds up the total time that the two have been touching and only causes the bell to ring when a certain level has been reached. In the prototype, using the component values shown, the total time that the two are allowed to touch can be varied from about 0.1 to 1.5 secs. This time range matches pretty closely average skills encountered by the author.

Most people, including children, can manage on the 1.5 sec setting (assuming that they try hard) but, as yet, no one has managed the shape used in the prototype on the shortest time.

The skill control is infinitely variable over the whole range, from 0·1 to 1·5 secs and this control is calibrated so you can even compete against your-

self, trying for a "best" result. On a sideshow, the control can be set to "easy" for children and to a shorter delay for adults, giving everyone a fair go. If you give a prize to anyone managing it and you find out you are giving too much away, the setting can be altered.

The circuit

Obviously no one would want to build in this extra sophistication if the cost was high but as you can

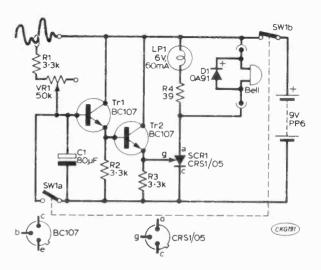


Fig. 1: The circuit of the Buzz-Bar.

see from the circuit in Fig. 1, it is very simple. Even assuming that you have to buy all the components brand new, this should set you back no more than £1. This is of course in addition to the cost of the simple unit; the costs of this with a bell, battery and a framework may come to another £1 or so.

The bar itself is connected to the positive supply line with the "wand" connected via R1 and a variable

part of VR1 to C1.

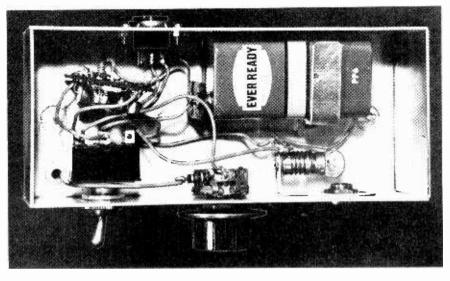
When the wand is touching the bar, Cl charges up through Rl and VRl. When the two are apart, Cl still holds the voltage. This is the main part of the circuit and controls the operation. When the voltage across Cl reaches a certain level, it is arranged that the bell should ring.

It is vital that we take only a tiny current from

such a small current that it can be considered unimportant.

C1 must be a first class quality component, able to hold its charge properly. Electrolytics vary enormously in their leakage and a decent one must be used. All British electrolytics tried have been found to be good, something which cannot be said for many imported types.

The emitter of Tr2 is connected to the gate of an SCR. As the voltage builds up on Cl, this also raises the voltage at the gate until the point is reached when the SCR switches on, causing the battery voltage to be applied across the bell. Two circuits are incorporated in conjunction with the SCR. It was found that the continual ringing of the bell was very annoying—especially indoors and for those



A underside view of the prototype. The various controls can be seen on the front panel. The circuit board is viewed from directly overhead. Compare this with Fig. 3.

C1 otherwise it would be discharging in between "touches." In other words we want to make sure that C1 holds its charge for a reasonably long time. For this reason we connect C1 to a high impedance stage, two emitter followers connected in series. This presents a very high impedance and draws

* components list

Resistors

R1 $3 \cdot 3k\Omega$

R2 3·3kΩ

R3 $3.3k\Omega$ All resistors $\frac{1}{4}W$, 5% types.

R4 39Ω

VR1 50k Ω lin. pot.

Semiconductors

Tr1 BC107—see text

Tr2 BC107—see text

SCR CRS1/05 (1amp, 50V)

D1 OA91

Miscellaneous

C1 80 μ F, 10V or similar, value is not critical; 6V, 60mA bulb; bulb holder; double pole toggle switch; battery clip; PP6 battery; planed timber; wire coat hanger; Veroboard; Aluminium chassis available from H. L. Smith Ltd, 287 Edgware Road, London W.2. Size: $6\frac{3}{8} \times 2\frac{3}{4} \times 1\frac{7}{8}$ in, price 60p including postage.

not actually doing the test. For this reason two indicators are used—a bulb as well as the bell; the latter can be disconnected. The bulb lights up when the "error" limit has been exceeded. If this facility is not required a resistor should still be wired in its place— 100Ω is about right. Since the bell is a make-and-break device the SCR will switch itself off as soon as the gate voltage falls, unless current is drawn all the time. Including this parallel resistor prevents premature switch off.

The diode across the bell prevents the build up of large back e.m.f. voltages which might otherwise

damage the SCR.

SW1 is a double pole switch. In the off position C1 is short circuited, ensuring that when the cycle starts that it is completely discharged; this short is broken when the unit is switched on.

The transistors shown are type BC107 but pretty well any silicon NPN types are suitable. In the unit shown these were later replaced by surplus devices, so salvaging good types for other uses.

Construction

There is nothing at all critical about the construction. In the prototype the electronic components were mounted on a small piece of Veroboard—the layout is shown in Fig. 2. No breaks are necessary in the conductor strip other than those around the mounting screw. This provides a neat and compact layout for the major components.

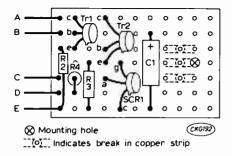


Fig. 2: The component layout on a small piece of Veroboard.

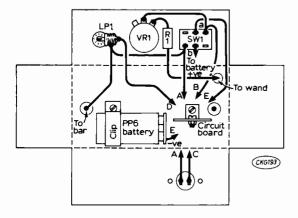


Fig. 3: The wiring diagram. The sides of the chassis have been bent out for clarity.

The prototype was built into an aluminium chassis available from the suppliers mentioned in the components list, but once again all sorts of arrangements may be used. Fig. 3 shows the wiring of the components not on the Veroboard; the wires to the board are marked A to E, matching up with the same letters in Fig. 2.

Two large holes are drilled in the top of the chassis to provide access to the bar itself. A piece of planed timber, lin×lin, is mounted on top of the chassis by means of two wood screws which also serve to hold the Veroboard mounting bracket and a battery retainer bracket.

The bell is mounted separately and the leads to this are connected via a plug. A DIN speaker plug and socket were used in the prototype but other types will do just as well.

One other hole is needed in the chassis; to bring the wand wire outside. This hole should be protected by a rubber grommet.

The bar

The bar itself is mounted by pushing each end through small holes drilled through the piece of timber. The holes should be quite small to make the fit tight.

Any firm wire can be used for the bar but a good source is a wire coat hanger; these are cheap and are often given away by dry cleaners. The part just below the hook, where the wires are twisted can be cut, these twisted wires will then fit nicely as a push fit into the body of a "Bic" ball-point pen

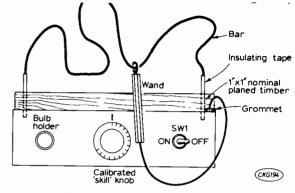


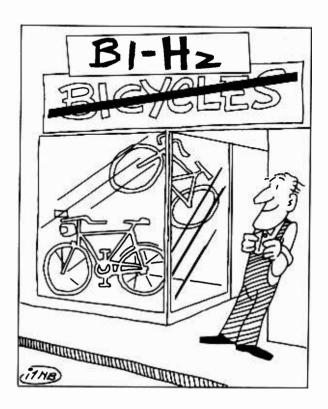
Fig. 4: Front view of the Buzz-Bar.

making a good wand. The wire should then be bent to form a tight loop.

Bands of insulating tape can be wound around the bases of both ends of the bar to prevent it ringing at the beginning and end.

Part of the fun will be bending the wire into shape. As we have said, the shape can be made virtually impossible and bends can be in three dimensions.

When someone does cause the bell to ring, the unit should be switched off. This will discharge the capacitor and allow another run to be tried. Many hours have been spent with the unit described here and although originally designed with children in mind, adults seem to monopolise it. It is only for fun but competitions and all sorts of games can be built up around it.



INNEXTMONTH'S

PRACTICAL WIRELESS



Whether your interests are short wave listening, medium wave DX'ing or listening to FM radio, you will find an aerial design on this wall chart that will improve your reception. Get rid of that odd bit of wire and use our design data to make an aerial for your favourite hand. Give that stereo tuner a chance to work properly by feeding it with a solid signal from a cecent aerial. Finally, medium wave DX'ers will really appreciate the data for a loop aerial that will effectively reduce that European QRM.

THE





PRE-TUNED CAR/PORTABLE RADIO

Using just six cheap transistors, this radio provides four pretuned switched channels, three of which can be adjusted to stations in the medium wave band, the fourth being intended for reception of Radio 2 on long waves. In the home or garden, the internal ferrite aerial is sufficient to provide a good signal in most areas. Placed on the parcel shelf of a car it will perform well utilising the car aerial, the pre-tuned facility being especially useful in moving vehicles.

BE CERTAIN

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THE NEXT ISSUE

PRICE 20p

logiprobe



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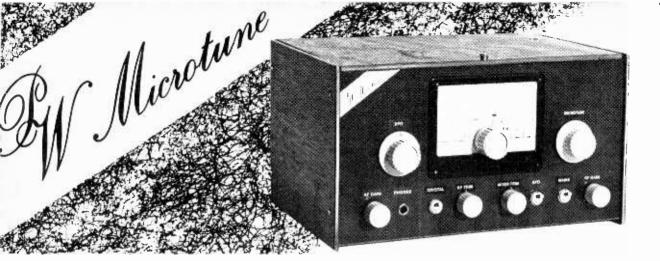
AUDIO REFERENCE SOURCE

What do you use an audio signal generator for? Certainly the full frequency range is useful but 95 per cent of the time spot frequencies are sufficient. This project describes a simple unit with square wave outputs at 100Hz, 1kHz and 10kHz with a calibrated output anywhere in the range 0. ImV to IV. The cost is only a fraction that of a conventional design and it makes an ideal unit for those who have only occasional use for an 2.f. signal source.



ALL IN THE OCTOBER ISSUE ON SALE 1st. SEPTEMBER

415



ITH congested amateur and short wave bands generally, there is more and more need for really sharp selectivity, and this circuit is excellent in this respect as well as being a relatively easy and inexpensive project. Coverage is approximatly 515kHz to 30MHz in four bands, and the use of a 1.6MHz intermediate frequency gives much better freedom from second channel interference on the high frequency bands, than is obtainable with a similar circuit using 465kHz i.f. The receiver is suitable for the reception of a.m., c.w. and s.s.b. signals.

Circuit

Fig. 1 is the circuit, and five valves provide the following functions: V1, r.f. amplifier (6BH6): V2, mixer and local oscillator (ECH81): V3, beat frequency oscillator (6C4): V4, i.f. amplifier and diode demodulator (EBF89): V5, a.f. amplifier and output stage (12AT7).

Aerial, mixer and oscillator coils L1, L2 and L3 are tuned by the 3-gang capacitor VC1/VC2/VC3. For maximum possible efficiency throughout all ranges, and with any aerial, the panel aerial trimmer VC4 is provided. VC5 is a similar mixer trimmer. To avoid band switching and the attendent losses L1. L2 and L3 are miniature plug-in coils.

For exact tuning of signals in a congested band, the "microtune" capacitor VC6 is used, having a panel control. This is also very useful when keeping a difficult signal in tune, where even a small movement of the main ganged tuning capacitor may lose it. VC6 also becomes essential when receiver selectivity is at maximum, despite the reduction drive on VC1-3.

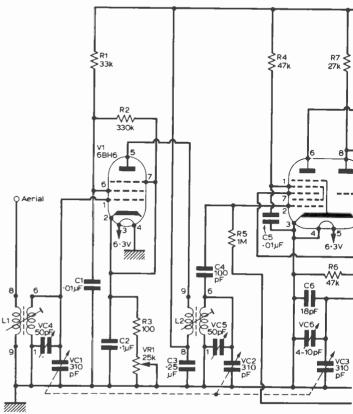
Two degrees of selectivity are provided. For normal reception S1 is closed. This gives easy tuning, and ample bandwidth for proper reception of broadcast or similar signals. With S1 open, and the crystal circuit correctly adjusted with TC1, selectivity is very much greater. Closing S2 brings in the b.f.o. necessary for c.w. and s.s.b. reception. No product type detector is included because strong s.s.b. signals can be reduced by the r.f. gain control VR1, while V3 is coupled to the grid circuit of V4 providing a suitable balance of s.s.b. signal and local carrier, so that the diode can function satisfactorily.

Fig. 1: Complete circuit of the Microtune receiver

V5 is a small double triode, used as a.f. amplifier and output stage. This is economical on h.t. and heater current while giving reasonably powerful loudspeaker reception. The secondary of T1 runs to a panel socket where phones or speaker may be plugged in. T2 is a small mains transformer providing up to about 40mA h.t. and 1.5A for the heaters. HT drain is around 30mA and heater current 1.2A, thus leaving the possibility of fitting an extra i.f. stage later, or using a triode-pentode in the V5 position for greater audio output

Chassis and cabinet

This is of novel construction, the result being inexpensive, easy to make and of quite satisfying appearance.



4-band receiver with variable selectivity

F. G. RAYER

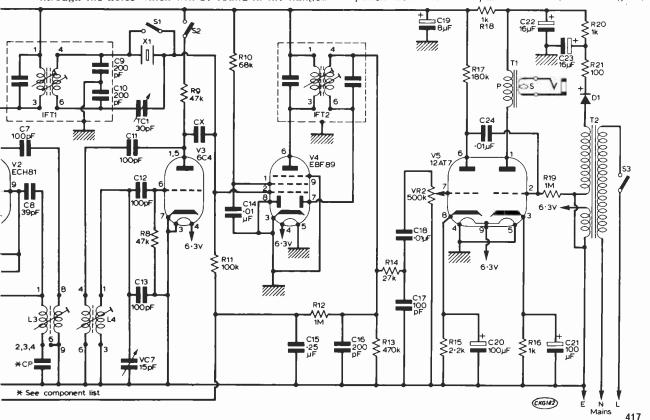
The chassis itself consists of two $12\times2in$, and two $8\times2in$, flanged "universal chassis" members with a $12\times8in$, flat plate. The same screws which hold the 2in, wide members to the plate also secure two $12\times4in$, and two $8\times4in$, flanged members on top, thus forming a complete chassis/cabinet $12\times8\times6in$, high. The final appearance of this cabinet, with the top closed by a hinged $12\times8in$, plate, is quite reasonable, and is obtained with the minimum amount of work.

In the receiver shown, it was decided to cover the front with a sheet of $_{16}^{1}$ in paxolin, held in place with component fixing nuts. Sides are cut from 3-ply wood and varnished. These are $8^{1}4 \times 6^{1}2$ in. to give a little overlap at top, bottom and front. The sides are held with chrome headed 4BA screws run through the holes which will be found in the flanged

members. A lid to match is also cut from plywood.

The best way to facilitate construction is first to bolt together the $12 \times 2in$. and $8 \times 2in$. members, placing the end flanges of the $12 \times 2in$. members outside the $8 \times 2in$. members. Check for squareness, put the $12 \times 8in$. plate on, and drill three holes along each 12in. edge, and two on each 8in. edge, about 1^12in . from the corners.

Drill through the chassis flanges to match, so that these can be bolted to the $12\times8in$. flat plate. Assemble the $12\times4in$. and $8\times4in$. members to match, put the plate on these, and drill the flanges. The whole can then be bolted together, though construction is easier if the $12\times4in$. back member is left off until later. In any case, drill and punch holes for the components as in Fig. 2 before assembly, and punch holes for the panel controls, as in Fig. 3,



before screwing on the 12×2in. front runner.

Mark out and drill the 12×4 in. front member before finally screwing it in place.

If an overall panel of thin paxolin or other material is to be fitted, mark it now by holding it in position and scribing through the existing holes with a sharply pointed tool. Holes will then match up correctly.

Above the chassis

Holes are drilled near T1 and T2 so that adequately insulated leads can pass down through the chassis. Fit valve and coil holders with tags as in Fig. 3, including soldering tags under the nuts. Drilling positions for L4 and the i.f.t.s can be found by pressing paper on the pins, holding this on the chassis, and marking through with a sharp point. A central hole is necessary to allow adjustment of the cores. It is as well to put pieces of insulated sleeving on the pins.

The front is carefuly marked with the height of the ganged capacitor spindle, so that the drive can be fitted, and the spindle should line up exactly with this. Solder insulated leads to the tags of VC1, VC2 and VC3 and pass them down through holes, before bolting the capacitor in place.

There is some latitude in the choice of VC6, and a component with a maximum value between 4pF and 10pF is most suitable. If VC6 is too small, it may give adequate tuning near the h.f. end of a band, but not near the l.f. band end. The capacitor should be smooth and electrically silent in action. Cheap surplus capacitors in good condition may be used by removing some plates.

In a similar way, though 15pF is shown for VC7, values of from about 10pF to 25pF are possible.

Below the chassis

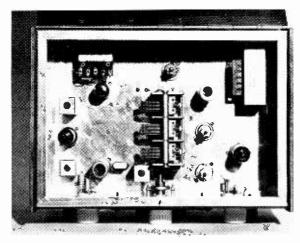
All connections etc. are shown in Fig. 3. Capacitors VC4 and VC5 need not be 50pF, but values of about 25pF to 50pF are most suitable. VC4 is on a bracket cut from scrap metal, and operated by a ¹4in. insulated or metal shaft and coupling.

The aerial socket is an insulated type and the adjacent terminal on the back runner is for earth.

The contact-cooled rectifier is bolted flat on the chassis side, and all burrs should be removed from the associated holes after drilling. Capacitors C22 and C23 are a double unit in Fig. 3, but separate capacitors give the same result. Tag strips are used to support various items and connections.

In general, run heater and h.t. leads against the chassis. Leads to r.f. and other circuits should be clear of the metal chassis. It is useful to employ differently coloured wires or sleeving, such as red for h.t. positive, blue for heaters, etc.

The leads to VR2 run along the side of the chassis as shown. All leads should be reasonably short and direct, especially those of grid and anode circuits, which should be well separated from each other. Leads run from the primary of the output transformer T1 to C22 positive, and pin 1 of V5. Leads from the secondary run against the chassis to the miniature output jack. The transformer is intended for a 3 ohm or similar speaker, which should be in a cabinet or fixed to a baffle board. Put extra sleeving over the primary leads of the mains transformer and bring them down to the "neutral" tag N, and to S3. Connect S3 back to the tag L of the tag strip. Bun a 3-core flexible cord from the tag strip, brown to L,



View into top of cabinet. Compare with Fig. 2, right.

blue to N, and green-yellow to E (metal chassis). Connect the cores correctly to a 3-pin plug fitted with a 3A fuse.

The 6.3V secondary leads pass through a hole, and are taken to MC and pin 4 of V1, Fig. 3. The h.t. leads go to MC and rectifier negative.

Modifying IFT 1

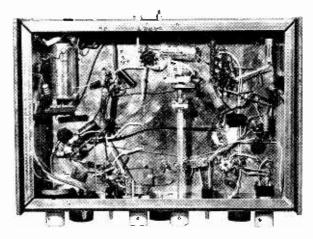
If wished, the receiver can be tested by fitting i.f.t.1 as supplied, omitting TC1 and the crystal, and wiring pin 4 to pin 2 on V4. However, i.f.t. has to be modified, to secure a balanced output circuit, when fitting the crystal.

Straighten the screening can tags and remove the can. 100pF capacitors are internally fitted between pins 1 and 3, and between pins 4 and 6. Cut the leads of the capacitor between pins 4 and 6, and remove it. Replace the can and bend over the tags.

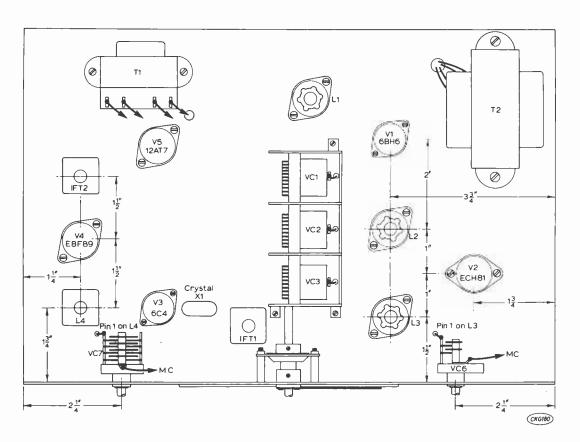
The i.f.t. is fitted as in Fig. 3, and C9 and C10 (each 200pF) are soldered from MC to pins 4 and 6, as shown. The receiver can be tested in this form without TCl and the crystal.

Crystal filter

The i.f.t.s are intended for 1.6MHz, but have some range of adjustment, so that it is not essential that

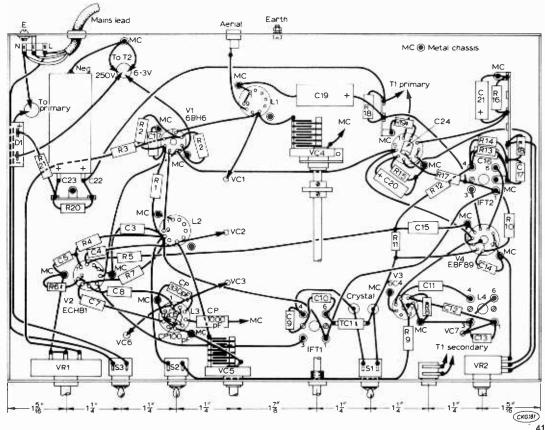


This view may help when wiring underneath the chassis from Fig. 3, right.



▲ Fig. 2: Layout of components mounted on top of the chassis.

Fig. 3: All wiring is shown in this under chassis diagram.



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the crystal be exactly this frequency. But to avoid the possibility that the i.f.t.'s cannot be adjusted to the crystal frequency, the latter ought if possible to be in the 1575-1625kHz range. In some commercial equipment, where an interfering transmission may fall around 1.6MHz, it is usual to fit a 1.6MHz wavetrap in the aerial circuit. In the receiver described, no trouble of this kind was encountered.

Adjustments to the crystal filter are best carried out with a signal generator, or, if this is not available, by tuning in a strong, stable signal, such as a

BBC transmission on medium waves.

With S1 open and TC1 about half closed tune slowly through a strong signal, to locate the crystal resonance. It will probably be found that the response of the i.f.t.'s gives strong reception over a few degrees of the tuning scale. Near this, one side or the other, the crystal resonance point should be heard. This may be quite weak if far from the i.f.t.'s in frequency, but should be very sharp, having a peak covering only a small part of a degree on the scale.

If this peak is not found, close S1, move all the i.f. cores a little one way, and search for it again. If it is still not found, repeat, moving the cores the

other way.

When a signal can be tuned in on the crystal resonant point, adjust all the i.f.t. cores to this, using a proper core adjusting tool. This should give an enormous increase in sensitivity and signal input should be reduced, or VR1 turned back. A high resistance voltmeter may be clipped across C14 (negative to chassis) and adjustments made to secure maximum voltage, as this is more accurate than adjusting by ear.

TCl should then be adjusted to balance stray capacitance. This is easy with a signal generator, but

can be done without it.

Tune slowly through a stable transmission, carefully noting volume (or preferably the voltmeter reading). It will probably be found that there is a dip in volume one side the signal, and a less selective response the other side. Adjust TCl a little at a time, noting how this changes. When TCl is adjusted too far, the dip or notch will move to the other side of the signal, and the less selective response will appear on the opposite side of the signal to that originally. The best setting for TCl is midway between these two situations. Tuning should then be extremely sharp, while music will sound very muffled, due to sideband clipping.

This type of crystal filter can be used with a variable capacitor instead of TC1, with panel control, but this has to be completely insulated from the chassis and panel. A variable control allows the filter rejection notch mentioned to be moved across the pass-band of the i.f. amplifier, reducing or elimin-

ating an interfering carrier.

With VC7 half open tune in accurately an a.m. transmission, close S2, and rotate the core of L4 until a strong heterodyne is heard. Set the core in the zero beat position and seal. Moving the core either way from this position causes a tone which rises in frequency. Rotating VC7 from the central position has the same result.

Cx is an extremely small coupling capacitance, and it may be found that the lead from R11 to pin 2, V4, is near enough to C11 to provide this. If not, run an insulated wire from pin 5, V3, near this lead.

If the coupling of V3 is too great, weak signals will be lost. If, on the other hand, coupling is very small,

★ components list

Resid	stors	•	
R1	33kΩ 1W	R8 47kΩ ‡W	R15 2·2kΩ ½W
R2	330kΩ 1W	R9 47kΩ 1W	R16 1kΩ ‡W
R3	100Ω 3W	R10 68kΩ ½W	R17 180kΩ ‡W
R4	47kΩ 1W	R11 100kΩ 1W	R18 1kΩ 1W
R5	1MΩ 1 W	R12 1MΩ 1W	R19 1MΩ ½W
R6	47kΩ ½W	R13 470kΩ ½W	R20 1kΩ 2W
R7	27kΩ 1W	R14 27kΩ 1W	R21 100Ω 1W
		-	

VR1 25k Ω pot. linear. VR2 500k Ω pot. log.

Capacitors	
C1 0·01μF 350V	C13 100pF SM
C2 0·1μF 150V	C14 0·01µF 350V
C3 0·25µF 350V	C15 0 25µF 150V
C4 100pF SM	C16 200pF
C5 0.01pF 350V	C17 100pF
C6 18pF SM	C18 0·01µF 150V
C7 100pF SM	C19 8µF 450V
C8 39pF SM	C20 100µF 6V
C9 200pF 5%SM	C21 100µF 6V
C10 200pF 5% SM	C22 16µF 450V
C11 100pF SM	C23 16µF 450V
C12 100pF SM	C24 0 01 uF 350V

Padding capacitors: Range 2, 100pF SM: Range 3, 330pF SM: Range 4, 1000pF SM.

VC1-2-3 3 x 310pF gang (Jackson-E type) VC4 50pF variable (type C804) VC5 50pF variable (type C804) VC6 5pF variable (type C804) VC7 15pF variable (type C804)

TC1 30pF compression trimmer

Valves

V4 EBF89 V1 6BH6 **V2 ECH81** V5 12AT7 V3 6C4

Inductors

L1 Miniature plug-in, valve type (Denco "Blue"). L2 Miniature plug-in, valve type (Denco "Yellow"). L3 Miniature plug-in, valve type (Denco "White"). (Frequency ranges as in text)

L4 BFO coil (Denco BFO2/1-6) IFT1/2 IF Transformers (Denco IFT16)

Metalwork

"Universal chassis" flanged sides: 12 x 4in. (2), 12 x 2in. (2), 8 x 4in. (2), 8 x 2in. (2), panels 12 x 8in. (2) (Home Radio).

Miscellaneous

T1, Valve output transformer (Home Radio TO43). T2, Mains transformer, 250V 40mA, 6.3V 1.5A. (Home Radio TM24A). X1, crystal 1.6MHz, type HC6U, and holder (Senator Crystals). D1, contact cooled rectifier, 250V 50mA. Valveholders, B7G (2, with screening cans), B9A (3). Dial and drive (Jackson 4103/A), coupling (Jackson 5610). On/off toggle switches (3). Headphone socket. Polystyrene rod and panel bush. Knobs.

s.s.b. signals, even when resolved, will sound like very much over-modulated a.m. However, VR1 allows signals to be reduced, so Cx is not critical.

When receiving s.s.b., turn VC7 one way or the other, as needed to resolve the signal. The a.f. gain (VR2) should usually be well advanced, while strong signals have to be reduced at the demodulator diode by turning back VR1. With correct adjustment, the circuit gives very good results in s.s.b.

CW is dealt with in a similar manner, VC7 adjusting the beat-frequency note. Switch S2 is off for a.m. reception, and VC7 is inoperative.

It will be seen that strong injection from V3 will create an a.g.c. voltage, reducing gain, but this was not found too important, using minimum coupling, as explained. This effect can if wished be overcome by switching the a.g.c. out for c.w. and s.s.b. reception.

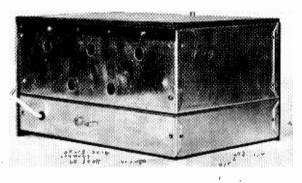
Band coverage

Coils are inserted in a set of three for each band, and the ranges are approximately as follows:

Range 2. 515-1545kHz. Range 3. 1·67-5·3MHz. Range 4. 5·0-15·0MHz. Range 5. 11·0-30·0MHz.

Range 2 gives m.w. coverage, and may not be required. Range 3 includes the two l.f. amateur bands 80 and 160m, and many other transmissions. Range 4 covers most general s.w. broadcasts, with Range 5 taking over for the h.f. bands.

Coverage is adjusted by closing VCl-3, and adjusting the core of L3 for the l.f. band end. To simplify adjustment, C6 is fixed, and no adjustment is made at the h.f. end of the bands. If this is required, replace C6 by a 30pF trimmer.



This not very inspiring view of the back of the receiver shows how the flanged "universal chassis" members are boiled together with a flat plate in between, forming the top of the chassis.

Tune in a signal near the h.f. end of the band (VC1-3 open) and peak VC4 and VC5 for best volume. Tune towards the l.f. end of the band, adjusting the cores of L1 and L2 to give best volume. When the cores of L1 and L2 are correctly adjusted, little movement of VC4 and VC5 will be required. However, some adjustment of these controls will improve weak signals. No efficiency is lost if both VC4 and VC5 can be peaked with any set of coils, and are not then fully open or fully closed. The core adjustments are thus not too critical.

L1 and L2 are separated and partially screened by the can on V1, and by the ganged capacitor. With some aerials it may be found that VC1 becomes regenerative at maximum setting of VR1 at some frequencies, and this effect can be used to boost sensitivity.

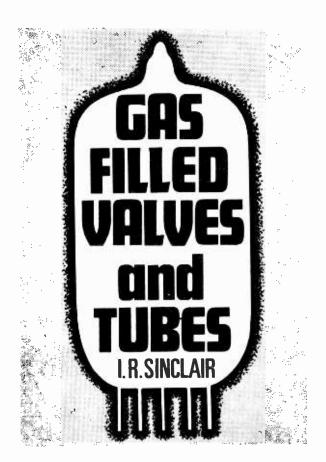


PROPAGATION over the long sea path between the UK and the northern part of South America is often good during the late summer. From midnight until sunrise is the time to search for medium wave stations from this area and many Latin Americans can be heard, Venezuelans being particularly prominent. Listen for Radio Giradot, Maracay on 650kHz; Radio Rumbos, Caracas on 670kHz; Radio Caracas 750kHz; Radio Puerto la Cruz 760kHz; Radio Margarita, La Asuncion 1020kHz; Radio Zulia, Maracaibo 1080kHz; Radio Carupano 1110kHz; Ordas del Lago, Maracaibo 1120kHz; Radio Tiempo, Caracas 1200kHz; Radio Valera 1230kHz.

Medium wave broadcasters in the Caribbean area are capable of being logged in the UK at this time of year. The most conspicuous is PJB on 800kHz which is a 525kW outlet of Trans World Radio, located on the island of Bonaire in the Netherlands West Indies. This station broadcasts religious programmes in English and other languages and it also relays Radio Nederland. Other English speaking transmissions to look for are from Georgetown, Guyana on 760kHz; Radio Belize on 834kHz; Radio Caribbean in St Lucia in the Windward Islands 840kHz; Radio Victoria, Arubia, NWI on 925kHz; ZDK St Johns, Antigua on 1100kHz.

DX'ers generally find it useful to have a list of known broadcasting stations arranged in order of frequency. J. S. Smith of Enfield, Middlesex; S. F. Hannaford, Plymouth and David Cotterall, Helensburgh, Scotland have enquired about published lists of medium wave stations. "Guide to Broadcasting Stations" Butterworth Press is on sale in many bookshops at 50p. It covers all medium wave stations in the European area and North Africa plus a number of the more powerful ones from North and South America and it should be of value to the newcomer to the band. For the serious DX'er there is the World Radio-TV Handbook published in Denmark and distributed in this country by Fountain Press. It costs £2.80 but in addition to comprehensive lists of the medium wave stations in the different continents it contains detailed information, by country, of radio stations and broadcasting organisations in every country in the world. For the specialist there is "Broadcasting Stations of the World, Part 2" produced by the Foreign Broadcasting Information Service of the United States Government. All known stations outside the United States in the range 150kHz to 26MHz are listed. The current edition, can be obtained for \$2, post paid, by writing to the Superintendant of Documents, Government Printing Office, Washington, DC 20402, USA quoting catalogue number PX EX 7.9.971 Part 2.

Please send logs and information about the medium waves to the author at 132 Segars Lane, Southport PR8 3JG.



N AN AGE of semiconductors, the continued and increasing use of valves may seem curious, the use of gas-filled valves even more so. The facts are that semiconductors have by no means completely conquered the electronics market; there are still useful jobs for which valves are either more suitable or considerably cheaper, and there are some jobs for which only gas-filled valves are suitable at

The conventional hard (meaning evacuated) valve uses a cathode, heated by a wire filament, as its source of electrons which will then travel from the cathode to any positive electrode in the valve. The current which can be drawn depends on the rate at which the cathode can emit electrons and the field which the anode can create to attract them. There is little difficulty in creating high fields, an anode at a higher voltage or placed nearer the cathode will ensure higher fields, but there is a limit to the current which can be taken from a cathode of any given area, and there is also a limit to the field strength which can be applied to the usual oxide cathode without pulling the coating off the nickel tube which supports it.

The effect of gas

Before we can understand the effect of gas in a valve, we have to understand the way in which a gas behaves at low pressure. Gases, like other materials, are made up of atoms which are usually grouped together as molecules. Air, for example, is mainly a mixture of two gases, oxygen and nitrogen, and each of these gases consists of molecules of two atoms each. Carbon dioxide, another gas present in air, consists of molecules containing three atoms, one of carbon and two of oxygen. The important difference which distinguishes gases from liquids and solids is that gas molecules are spaced well apart; the average spacing at normal pressure is about 300 times the diameter of the molecules, so that the molecules are fairly free to move about and exert force (noticeable as pressure) on the walls of the container. If a small amount of gas is allowed to enter a large evacuated container, the distance between the molecules increases and the gas fills the container; we say that the gas has expanded. If we compress a gas to fill a small container, the spacing between the molecules decreases; we can decrease it so much for some gases that they turn into liquids. The laws connecting the pressure and volume of a gas were discovered by Boyle some 300 years ago, but it was only in the middle of the last century that it was discovered that gases at low pressures conducted electricity.

Electricity is conducted when charged particles move in a substance. In metals, the free electrons are the charged particles, in liquids, mainly solutions of certain materials in water, conduction takes place by positive and negative particles together. These particles are termed Ions; they are charged fragments of atoms which have been split by the action of dissolving the solid in water. In gases, we have molecules moving at high speeds around 500m per second at normal room pressure and temperature) and colliding with each other violently, but all this movement does not cause conduction because the particles are not charged. We can make charged particles of the molecules only by the violent means of splitting the molecules into atoms and the atoms into charged ions, as in the case of liquid conductors. We can do this by hitting the molecules so hard that they separate into atoms and the atoms lose one or more of the electrons which surround the central portion. We are then left with large positive ions and negative electrons. Note that these positive ions have a real independent existence, and can be collected, they must not be confused with semiconductor holes, which, though just as "real" in the sense of having measurable mass, velocity and charge, cannot be separated from the crystal lattice of the semiconductor.

lonisation

The splitting up, or ionisation, of the atoms of a gas can be carried out by any form of energy sufficient in size. Early workers with static electricity had noticed their capacitors rapidly discharging when they were held near a flame; the heat was producing some ions which were then attracted to the negative end of the capacitor while electrons, usually attached to atoms, arrived at the positive end. Ultra-violet light is another source of the energy needed to ionise a gas; Heinrich Hertz had noticed in his pioneer work on radio transmission and reception that a spark passed more easily between two electrodes at a given voltage if they were illuminated. The application of heat also causes ionisation; a hot ionised gas is called a "plasma" and conducts well enough to be used as a conductor at temperatures where metals would melt. Since heating a gas is simply a method of making the molecules move faster, we might expect that any method of making the molecules move faster might also cause the gas to become ionised if we apply enough energy.

If gases are so easily ionised, why do they not conduct electricity well? The answer is that ions in a gas cannot move easily to the anode or cathode, whichever is attracting them. Imagine a gas confined between two electrodes (Fig. 1) and one atom ionised. Clearly the positive ion will be attracted directly to the negative electrode and the negative ion to the

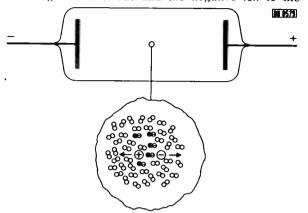


Fig. 1: The electrodes attract the ions but other atoms hinder this movement. The shape, size and spacing of atoms can be measured to a considerable degree of accuracy.

positive electrode; the forces act in straight lines from electrode to ion. It is, however, impossible for the ions to move in straight lines without colliding with other atoms, and it is the process of making collisions which accounts for the odd behaviour of gases. If the forces acting on the ions in our example are not very great, then the ions may rebound into each other and recombine, or they may slowly make their way, with many collisions, to the electrodes. The current in the second case will be so small that only electrostatic instruments will detect it. If we increase the voltage between the electrodes, still keeping the gas at normal pressure, then the forces on the ions will be greater and so the collisions will be more violent. Eventually, as we raise the voltage, the collisions may become so violent that new ions are created on each collision. When this happens, the number of ions available to carry the current increases enormously, and the current is large. We see a spark pass when this happens. If the voltage supply is well regulated, a continuous spark will pass; if, as is usually the case, the voltage drops when the spark passes, the ions are collected by the electrodes leaving no more available until the voltage rises high enough again to start more ionisation by collision, so that sparking is intermittent.

Reducing pressure

When we reduce the pressure of a gas, we are separating the atoms so that there are fewer collisions. Note that "fewer" still means many millions of collisions per second, even for currents of a few nanoamps. If we create a pair of ions in a gas at low pressure (about one ten thousandths of the normal atmospheric pressure) they will move very much further before colliding with atoms. Since a steady force of attraction between ion and electrode causes the ion to accelerate steadily, the ion can be travelling at a very high speed when it eventually hits another atom and it is very much more likely that it will cause the atom to ionise. If the pressure is low enough, each ion is capable of creating another ion pair at each collision, so that the number of ions

formed rises extremely rapidly. This sort of thing is called a "chain reaction" (Fig. 2) and it causes the gas to become a good conductor very rapidly. There are several important points about this chain reaction. It takes place only in a range of pressures; at high pressures the molecules are too close to allow the ions to reach a speed fast enough for ionisation by collision. Only if the voltage is made very high, about 30kV per inch gap between the electrodes, will ionisation take place to allow sparking. If the pressure is very low it may be possible for an ion to travel, on average, all the way from one electrode to the other without hitting another particle, in which case the current is only that carried by the original ions and is very low. Only in the range of pressures between these extremes can ionisation be reliably produced with low voltages (of the order of 150-300V) but high currents.

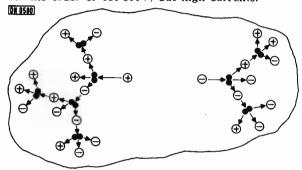


Fig. 2: A chain reaction occurs when ions can move far enough to accelerate to high speeds; when collision occurs more ions are produced.

The voltage used is also critical. Below a voltage called the "striking voltage," no ionisation takes place to a sufficient extent to cause noticeable currents because the voltage cannot give any ions which are formed enough speed to ionise other atoms by collision. The value of this striking voltage depends on the gas used, its pressure, and the materials used for the anode and cathode.

When current is flowing and the gas is already ionised, the voltage drop between anode and cathode is smaller than the striking voltage, and is called the "running voltage." Most of the voltage drop takes place near the cathode, as this is where most of the collisions take place due to electrons leaving the cathode. This region is visible because the excess ions recombine, causing light to be emitted. The value of the running voltage is also dependent on the gas used, its pressure and on the anode and cathode materials. In the running condition, the gas discharge behaves as a negative resistor, meaning

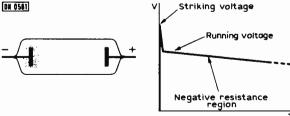


Fig. 3: A gas discharge appears around the cathode at fairly low pressures (about one millionth of atmospheric pressure). On the right is shown the characteristic for a gas tube.

that the voltage between the electrodes decreases as the current increases see Fig. 3. Contrast the behaviour of a normal resistor, where the voltage drop increases as the current through it is increased.

Armed now with some knowledge of the complex processes which take place in a gas at low pressure, let us now see how these processes can be used.

Gas filled rectifiers

An ideal rectifier should have a very small forward voltage drop when conducting, a very high reverse resistance, the ability to withstand high reverse voltages and the ability to pass high forward currents. For low voltages, semiconductors which answer this specification fairly well have been available for some time and have superceded the diode valves formerly used. Problems arise, however, when voltages of several kV and currents of several amps are required, as they often are in a transmitter. Connecting semiconductor rectifiers in series is not always suitable (it is difficult to ensure that the same voltage appears across each rectifier) and can be expensive when really high voltages are used. Valve rectifiers using oxide cathodes can be used up to a few kV, but the high electric field causes the oxide to peel off at higher voltages. Valves using thoriated tungsten filaments can be used, but the currents available are restricted, since these filaments are not such efficient producers of electrons as oxide cathodes. The solution, which has been used for many years, is the gas-filled rectifier.

If a rectifier contains some gas, then the electrons which are emitted cause ionisation of this gas, and a low anode-to-cathode voltage will be able to cause conduction. The resistance of the valve is very low, a combination of the resistance of the leads and the resistance of the gas, and the slope resistance is negative, as previously explained.

Since electrons are provided in large numbers by the cathode the running voltage is very low, but when the voltage reverses, electrons are no longer emitted from the cathode and the gas will ionise only if the voltage between the anode and the cathode exceeds the striking voltage. By making the pressure of the gas low enough, this striking voltage can be made almost as high as we like so that a gas rectifier can be used up to very large values of peak reverse voltage, several hundred kV. The valve-makers' problem is to make the pressure of gas constant despite the absorption of gas by the anode and the emission of other gases by the cathode. If the design of the rectifier is such that these problems are overcome, then the gas rectifier is reliable and long-lived. Note the difference between this case and that of the hard valve rectifier which has become gassy through leakage or overheating. Here there is no control of gas type or pressure, and the valve fails when the striking voltage is less than the back reverse voltage. In commercially made gas rectifiers, the gases commonly used are Mercury-Vapour for the smaller types, and Xenon for the larger varieties.

Some precautions are needed when using gas rectifiers. When mercury vapour rectifiers are in use, the temperature of the valve must be allowed to come up to full operating temperature before high voltages are applied. This is because the mercury is in the liquid state at room temperatures and has a very low pressure at such temperature. If high voltages are applied in these conditions, there is a risk of sparking between anode and cathode, or at least of a very high resistance discharge. Either way, the resulting dissipation can be very destructive to the cathode, and dangerous to the operator if the valve should

fracture, as mercury-vapour is poisonous. Most mercury-vapour rectifier circuits incorporate some sort of automatic delayed switching to ensure that the heaters have been on for at least one minute before the anode volts are switched on. Such a circuit is shown in Fig. 4.

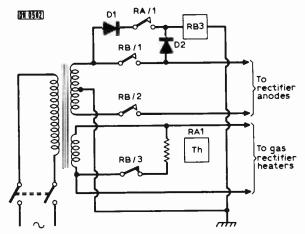
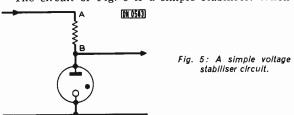


Fig. 4: A delayed switching circuit for gas filled rectifiers. The thermal relay in the heater circuit ensures that there is a delay before the h.t. is applied.

Voltage stabilisers

Another use of gas-filled valves is bound up with the stability of the running voltage. We have said earlier that, as current through a gas-filled rectifier increases, the voltage across the valve decreases. This change of voltage with current is small, and, with suitable choice of materials for anode, cathode, and gas, can be made very small. This is the essence of a voltage stabiliser, that a voltage should be unchanged while current varies.

The circuit of Fig. 5 is a simple stabiliser. When



voltage is applied at point A and raised until it equals the striking voltage of the gas, current will flow in the valve, and the voltage at B will be the running voltage. As the voltage at A is raised, the current through the resistor and the valve increases, but the voltage at B remains almost constant. If current is taken from B, it is taken at the expense of the current in the valve. If the current taken at B is greater than the current which was flowing in the valve, then the gas discharge will cease and the voltage at B will no longer be stabilised. The actual value of the running voltage, as said before, depends on the gas, and the materials used for anode and cathode; standard values obtainable with well-tried combinations of materials are 45V, 70V, 85V, 105V, and 150V. By more careful choice of materials, the change of voltage for a given change of current can be made smaller than normal, and the change of running voltage with temperature change can also be made very small. Such carefully designed

valves are known as voltage reference valves, and can be used, not in the simple circuit of Fig. 5, but in more complex stabiliser circuits such as that of Fig. 6, where the current through the reference valve is fairly constant.

Normally, the voltage difference between striking and running is fairly large, so that a 105V stabiliser needs about 150V to strike, and the voltage required for striking becomes higher when the valve is in darkness. The reason for this is that light provides enough energy to cause some ionisation in the gas, and so provide the conditions for easier striking. When light is excluded, the voltage must be raised to help start ionisation off from the very few ion

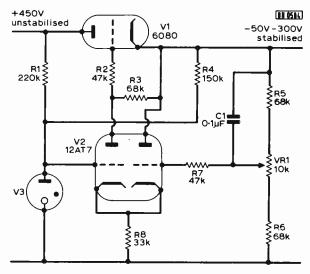


Fig. 6: A complex stabilised power supply.

pairs produced solely as a result of the temperature of the gas. To avoid the difference in striking voltages with light, which can be troublesome in reference tubes, the gas in the valve is often made slightly radioactive. The ions produced by radioactive disintegration then start off the ionisation in the valve, so that the striking voltage is much less dependent on the effect of light.

Neon lamps

Small neon lamps are commonly used as indicators of voltages greater than their striking voltage. Such lamps are used with a large resistance, in the order of $500k\Omega$, in series so that the running current is extremely low. There are, however, other uses. Since a neon is a gas-filled tube, the running voltage is fairly constant, and neons can be used as stabilisers for circuits which take low currents and for which a large degree of stabilisation is not necessary. For this purpose, neons have to be separated from the

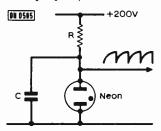


Fig. 7: A simple sawtooth generator. The capacitor must have low leakage and should not be an electrolytic.

series resistor which is built into the base of many types, or neons without a resistor purchased.

There is usually a large voltage gap between striking voltage and running voltage, and this can be

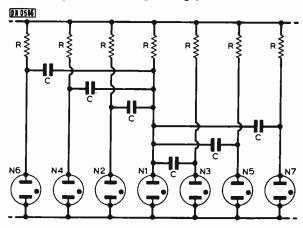


Fig. 8: A random flasher circuit, R can be between 220k Ω and 2-2M Ω and C between 0·1μF and 1μF (low leakage types must be used).

used in sawtooth circuits of the type in Fig. 7. Such signals are used as wide-range frequency sources, or as time delays where high accuracy is not needed. One amusing circuit based on neons is the random flasher of Fig. 8. When the supply is connected, one neon will inevitably strike before any other. The drop in voltage caused by the change from striking voltage to running voltage is communicated to all the others by the capacitors, and delays the rise of volts across all the other neons. The next neon to strike will have the effect of extinguishing the first and also of delaying the firing of the others vet again. The neons will then wink in random order as long as voltage is applied. This can be used as a decoration, but has serious uses when a large number of neons are used, as it can generate random numbers—the **ERNIE** principle is based on such a scheme for selecting Premium Bond winning numbers.

TO BE CONTINUED

CQ! CQ! CQ! CQ! CQ!

ISSUES WANTED

...The issue containing details of the 6V car radio.—P. J. Day, 43 Highgate Lane, Farnborough, Hants.

Farnborough, Hants.

The Issue containing users of the Treasure Tracer.—C. Weaver, 30 Cherry Garden Road, Canterbury, Kent.

The June Issue containing the Electronic Ignition System—S. Bennion, 312 Warrington Road, Galzenbury, Nr. Warrington, Lancs.

Jan, June, Aug, and Sept. 1970.—P. D. Williams, 52 Acacia Road, Sutherland, N.S.W. 2232, Australia.

The Issues containing Parts 1 and 2 of Transistor Circuitry for Beginners.—P. M. Bontomme, 26 Greencroft Gardens, Reading, Berkshire.

The issue containing the Treasure Tracer.—M. A. Branford, Flat "C", The Clint, 4 Parade, Chudleigh, South Devon.

October and November 1971 P.W.—R. Burke, 62 Garner Road, Walthamstow, London, E.17.

The issues of P.W. containing the three transieter ameliance.

...October and November 1971 P.W.—R. Burke, 52 Garner Roau, vyaurummeter, London, E.17.
...The issues of P.W. containing the three transistor amplifier Take 20 project.—P. Mieszkowsk, 25 Kingsway, Wembley, Middlesex,
...June 1964 P.W.—H. Hallybone, 7 Sauncy Avenue, Harpenden, Herts,
...September-December 1961 issues of P.W.—R. B. Howard, 3 St. George's Place,
Macclesfield, Cheshire.
...July 1917.—D. C. Dick, 97 Curtis Avenue, Kings Park, Glasgow, S.4.—
...July 1968 P.W.—P. J. Chapman, 90 Melody Road, Wandsworth, London, S.W.18.
...Fault-finding chart No. 1 issued April 1968.—H. Symonds, 16 Newhaven Street,
Piallea 4655, Queensland, Australia.
...April 1969 and subsequent copies of the P.W. Double-12 amplifier.—K. Stean,
101 Atherley Road, Shirley, Southampton, Hants.
...The issue of P.W. containing the 7MMŁ Transcelver,—P. Matlock, 60 Downs,
Avenue, Whitstable, Kent.
...February 1971 issues of P.W.—Mrs. M. M. Buckner, 13 Tankerton Road, Tankerton
Whitstable, Kent.

Avenue, ventissaure, rent.
...February 1971 issue of P. W.—Mrs. M. M. Buckner, 13 Tankerton Road, Tankerto
Whitstable, Kent.
...The Issue containing the Treasure Tracer.—R. Birkby, 1 Gloucester Place, Haughton Road, Darlington, Co. Durham.



HIS XYL never complains about the junk in her husband's shack, for the simple reason that he might one day take the notion to have a prowl round her shack and decide that it is full of junk which rightfully should be gracing the local rubbish tip.

I must admit that my husband's obsession has added somewhat to my vast store of bits of half-digested knowledge, useful or otherwise. Prior to my marriage six years ago I knew nothing about wireless beyond how to switch the set on and off, find a suitable station, and adjust the volume. An aesthetically satisfying array of lit valves might very well have been left-overs from last year's Christmas tree, and I wouldn't have recognised a silicon epitaxial planar transistor if one had jumped up and yelled, 'Boo!' I still wouldn't, but I read it in an advertisement and it sounds fascinating. Anything to do with transistorised planarians, I wonder?

"No, it isn't!" shouts a voice through the intercom connecting the two shacks. "And how about making some coffee?" This, despite the fact that his junk resides in the kitchen while mine is in a bedroom.

I comply with his request and brew the coffee while watching his usual antics as he tangles himself in various cables and wishes desperately that he could get a third hand for Christmas or increase the length of his arm because the part he wants is just out of reach and where the hell is the soldering bo . . . Aaah! "Found it, darling?"

Having consoled him with coffee and a kiss, I leave him muttering about microhenrys and picro-

farads. I sometimes have nightmares about poor Messrs. Henry, Faraday, Volta and all the others, wandering around in that maze of liquorice-allsort components.

In my own shack I am faced with nothing more frightening than a typewriter, a small sheaf of completed manuscript on the right, and a whacking great pile of unsullied, virginal white paper on the left. I have only one problem: that of intelligently and successfully transferring the reams of paper from left to right via the typewriter. My husband maintains that the simple and logical solution would be to shift it straight into the wastepaper basket.

I really can't imagine why Wireless Widows let themselves get into such a situation in the first place. They don't know what they're missing. Listening to shortwave radio can be, among other things, very entertaining, but even more so is watching one's husband listening to shortwave radio.

Having finished my "scribbling" for the evening, I make the cocoa, which comes very shortly after the coffee. I hand his mug to him, then retreat to a safe distance, that is, a distance at which I can watch yet not be seen to be watching.

There he sits with the earphones on his head, and the oddest expressions darting across his face. His lips purse, his beard wiggles, his eyebrows flit up and down like a couple of frenzied pussmoths. A calm swig of cocoa erupts into an unexplained gale of choking laughter, he scribbles something on a pad and mumbles a few indistinguishable words. I say, "Yes, dear," just in case they were intended for me. When he stands up suddenly to reach for his list of call signs he unknowingly pulls the jack-plug out and has a silent fit, trying to discover what has gone wrong with his beloved B40. Finally he finds the plug dangling around his knees, thrusts it back in, and performs a magnificent impromptu Indian war-dance because, during his frantic fiddling with the controls, he turned the volume full on and now doesn't know which to do first-turn down the volume, tear the headset off, pull the jack-plug out again, or simply collapse in a twitching heap on the floor!

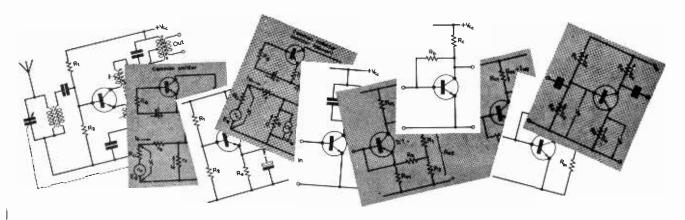
Oh, yes, ladies. Shortwave radio can be fun.

Then there's the "helving husband in the shack" act. Any XYL can do this, no previous experience required.

When he has spent half an hour searching for a certain component which he "had in his hand not two minutes ago", and it's the only one of that kind that he has, and the shops are closed so he can't get another till Monday; that is the moment at which the loving, devoted and helpful XYL enters the hallowed precincts and retrieves said component from floor/tobacco tin/inside of radio chassis/shirt pocket/hip pocket/coffee mug. The XYL then places the component quietly, unobtrusively and without comment, on the bench and beats a strategic retreat.

She is also handy for testing electrolytic capacitors, to see if they are safe or not. "Just touch this, darling, and tell me what you feel." Or for holding a tinv part in a restricted space while the expert solders blindly. And heaven help her if she is weak-willed enough to allow the "slight warmth" at her finger-tips to force her to relinquish her grip, thereby breaking the delicate joint.

Painful? I wouldn't ever want to do away with the radio shack. It keeps him quiet for hours, so what are a few burnt fingers?



TRANSISTOR CIRCUITRY for beginners PART 10 H.W. HELLYER & MICHAEL HOLLIER

Matching

It is no great hardship to answer letters that begin: "Thank you for your series of articles . . ." and go on to pose a query. More difficult, if only because one is obliged to clench one's teeth and answer as objectively as possible, when someone with an opinion to air uses a criticism of the previous text as an excuse to ride his own hobbyhorse. But, answered they must be, so let's begin with the pleasant task of selecting a letter from Mr. Edward Tarrant of Rochester, as representative of one body of readers, who ask questions on matching.

It has been said that audio engineers would be out of a job if their customers understood the simple rules of matching. They have been expounded before in these pages. In Part 4, for example, dealing with Buffer Amplifiers, I took some trouble to explain some of the difficulties and expound the basic "rule of ten".

To come back to our correspondent: he says that his trouble is one of adding together 'bits of circuits.' He has made up the Fig. 4 complementary pair of the 'Experimenter's Circuit Supplement' of PW January, 1972 and obtained highly satisfactory results and now wants to add the active tone control circuit, Fig. 102, Page 131 of the Mullard 'Transistor Audio and Radio Circuits', but is worried because the output impedance of the latter is only 180Ω . 'Can one follow it with a common-base stage?' he asks.

Well, yes, one can-but why should we go to that trouble? Input impedances of the PW circuits are in the region of $30,000\Omega$ and over. Just as the amplifiers we have described, and have built for this article. sport a low output and high input impedance. That, Mr.T. is the whole point. Basically, one can match a low impedance output into a high impedance input of the succeeding section of equipment without too much bother.

There are exceptions, true; no use sticking the very low Z of the mains supply into the very high Z of your crystal microphone socket unless you want your tape recorder to go up in smoke. But, if it is

millivolts to millivolts, i.e., small signal handling devices, then the 'rule of ten' can be invoked more often than not. This is to say that an output will safely match into an input which has an impedance of ten times or more than itself. So our $10k\Omega$ output wants to 'see' at least 100kΩ. And, in Mr. Tarrant's case, 1800 is very well catered for by upwards of $30k\Omega$.

It is not, repeat NOT, a matching rule that like equals like—that a $5k\Omega$ output has to be matched into a $5k\Omega$ input. If you stop and think about it, those conditions will cut the available voltage exactly in half. And, as one of the ideas behind this series was to produce those handy buffer-links that allow us to squeeze the utmost from a source that may be delivering barely enough output, then you can see what a mistake such 'exact' matching would be.

To take a practical example: loudspeaker matching. An amplifier is rated at X watts into 8Ω . So it wants to 'see' an 8Ω nominal load to produce that power. But take a close look at an output circuit and you will see that its actual impedance is more like a fraction of an ohm. In fact, one of the designer's aims is to get it as low as possible to derive the greatest power and efficiency.

So to repeat: match up! And if, like Mr. Tarrant, you are 72 years old, then I salute you, and hope that I shall still be enjoying the thrills of construction when my three-score years and ten have been rung up!

Adding stages

So, for that matter, does Mike, who has been sweating away over a tepid soldering iron, producing a direct-coupled circuit to illustrate this month's argument. He has a little farther to go to reach the Biblical milestone, but, wise beyond his years, suggests we explain at this stage what we are getting at. 'You know, I know, and perhaps the Editor may know,' he says, but the series of articles has entered its decade and we are still scratching around at fundamentals, so let's explain why!

It is possible to take someone else's evolved circuit and build around it, taking for granted the basic design work, and come up with a beautiful piece of electronic wizardry that all your friends will admire.

This is, fundamentally, what our other typical correspondent, Mr. E. F. Good, from Malvern, has complained about. In Part 7 of this series we described the Darlington Pair transistor, and gave some of the calculations from which it was derived. This, we thought, was a legitimate demonstration of the evolution of transistor circuitry, from which the beginner could see where the application of the rules we had laid down might be leading him. The insertion was a legitimate 'leader'—but Mr. Good took exception to it.

'You do let your enthusiasm carry you away,' he complained. The only place where he would be likely to use it would be as part of the triples of a power amplifier. In other positions, a pair of opposite conductivity would be more suitable and would cost only a little more. 'A manufacturer's economics are different.'

Well, of course, Mr. G., but give us credit for some insight: we are coming to the complementary pair. All in good time. This series is for the beginner, which your letter shows you not to be. Palpably, for you complain that our circuit that was intended to get the best from a crystal microphone could have been better served by a j.f.e.t.

That is true, but, I beg leave to argue, irrelevant. Until we have got the hang of simple circuitry, there is no point in progressing to more sophisticated devices. Much of the rest of this long letter, picking out points from preceding articles—in no sequential manner, either—has been dealt with in Part 9. Some wrongly captioned drawings have been explained. Some omitted references (for simplicity) have been taken up, and there is one formula in which no account was taken of r₆, because we did not at that point want to confuse beginners: we have since dealt with that subject.

Mr. Good ends his letter with the statement: '. . . instructing beginners is the most responsible job one can undertake, and should be performed with the greatest possible care.'

Agreed, Sir, so without more ado we shall go on to discuss the biasing of BC109 transistors, with particular reference to the modifications necessary in our thinking when we add two stages together.

Coupling up

It is possible, in theory, to cascade similar circuits like erected chains of dominoes. But if we do, taking no account of voltage swings, etc., our amplifying edifice is likely to fall down. When this series was being discussed, the Editor wanted us to call it 'Building Bricks', and this example is a case in point. We must consider the composite circuit, the two transistor unit, when we add one stage to another.

Take Fig. 52. Here, we have Trl operating as a collector follower, capacitively coupled via Co to the second transistor, Tr2, working as an emitter follower. Bias for Trl is provided by a conventional potential divider, R1 and R2.

The input signal is coupled to base of Trl by $C_{\rm in}$. A measure of gain may be expected from this stage in its collector follower mode. The emitter resistance $R_{\rm el}$ is bypassed by $C_{\rm e}$ and the amplified signal at the

collector of Trl is capacitively coupled to the base of the following stage. All very good, except that, as some of our corespondents have been quick to point out, the output impedance is higher than would be desirable. So, adopting Mr. Tarrant's suggestion, almost, we use another stage between our Trl and whatever we are feeding.

This time, we are not interested so much in the gain—we've got that—but must preserve what we have got and present a low output impedance. So the second stage is operating at slightly less than unity gain in its emitter follower mode, but it does give us a low output impedance. The output voltage is developed across Re2 and is coupled to the following stages via Cout. The snag? The circuit, as given, is wasteful of components. It is a general rule that extra components in any circuit are extra possible sources of trouble: noise, instability, excessive current drain. So we shall proceed to whittle down the circuit.

To begin with, there is an alternative way of biasing Tr2. Readers of the previous articles will have no trouble in following the train of thought. We have already dealt with the collector follower and the emitter follower circuits. From the information given in past articles, the 'circuit' of Fig. 52 could be derived without much trouble.

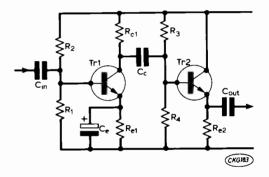


Fig. 52: Two stages already considered, a common emitter (collector follower) stage with an emitter follower stage added to it. The object being to produce a low impedance output while preserving the gain. This method of adding stages is wasteful of components.

Our alternative method is to remove bias resistors R3 and R4 and the coupling capacitor, $C_{\rm c}$. We now couple direct: that is to say, the collector of the first stage is taken directly to the base of the second. The result is that the base bias for Tr2 can now be derived from the collector of Tr1. Hence the term 'directly coupled.'

Operating conditions

Calculating the operating conditions is a matter of logical progression, as before. We start with Tr2. We know the transistor characteristics—we have dealt with the BC109 before. We know the supply voltage and the 'rail' voltage, Vcc, which we shall adjust to 8 volts by manipulation of the series resistor. Rdec, through which all the current in the circuit flows, Fig. 53.

The things we do not yet know, and which we shall have to define, are the load at the output and the drive voltage at the input.

Remembering what we have said about matching already, we shall define the 'worst case' load. If we decide that the maximum r.m.s. output voltage we require is, say, 200mV ($0\cdot2\text{V}$), and that the *minimum* impedance into which our circuit can feed is $5k\Omega$, we have some more guidelines to follow. Next thing we want to know is the peak voltage—remember, it is the voltage swing that gives us the operating voltages, i.e., defines our limits.

Here, we have $0.2V \times 1.414$ (Peak=r.m.s. $\times \sqrt{2}$)=

0.2828V peak.

For a 300mV r.m.s. output, we shall need $0.3 \times 1.414 = 0.4242V$ peak, and this is the value we shall choose.

To calculate the current which will be driven into our minimum load of $5k\Omega$, we divide our peak voltage by the load impedance,

$$\frac{0.4242}{5000} = \frac{4242}{50} = 82\mu A$$

To allow for a safe margin, we should operate Tr2 at an emitter current of five times or more than this. A convenient emitter current to use in calculations would be 1mA.

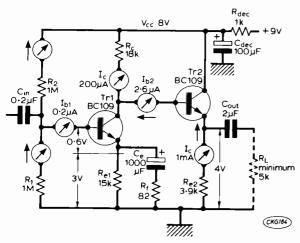


Fig. 53: A practical direct-coupled circuit. Linearity is good and distortion very low.

Operating Tr2 at an emitter current of 1mA and with an emitter voltage of 4V (half $V_{\rm cc}$, which is 8V), giving us a swing of 8V, peak-to-peak, again a good margin of safety is obtained.

Using Ohms Law, we calculate the value of Re2.

$$R = \frac{V}{I} \text{ or } \frac{4V}{ImA} = \frac{4000}{I} = 4,000\Omega$$

The nearest preferred value of resistor will be $3.9k\Omega$.

From the graphs

Turning now to the base conditions of Tr2. We have already said that bias is derived from the collector of Tr1 (see Fig. 53). We know that a silicon transistor such as the BC109, when forward biased, will have a voltage of 0.6V developed across the base-emitter junction. If we add this to the emitter voltage of Tr2, we get an expected base voltage of 4.6V, and so can say that the collector voltage of Tr1 will be the same, since the two transistors are directly coupled.

To calculate the base current of Tr2, we need to refer to the graphs which have previously been

published, remembering that $I_b = \frac{I_c}{h_{FE}}$

We know that, for a BC109, the $h_{\rm FE}$ (d.c. current gain) at a collector current of 1mA is 380. We know the collector current, and can now say:

$$I_b = \frac{1mA}{380} = \frac{100\mu A}{38} = 2.6\mu A$$

The collector current of Trl should be four or five times greater than this. To get a more satisfactory $h_{\rm fe}$ (a.c. current gain), we shall choose a collector current for the first transistor of $200\mu A$. Again, plenty in hand. In comparison with the base current of Tr2, this collector current of Tr1 is very large, and the effect of $I_{\rm b2}$ on the load of Tr1 can be ignored in our calculations.

This load,
$$R_c = \frac{V_{cc} - V_{cl}}{I_{cl}}$$
 or
$$R_c = \frac{8-4.6V}{200 (\mu A)} = \frac{3.4}{200 (\mu A)} = \frac{34,000}{2} = 17,000\Omega$$

The nearest preferred value is $18k\Omega$.

Emitter voltage

Referring back to Part 9, a table was given to show the effects of a spread of emitter current, resulting from changes in $H_{\rm FE}$ and $V_{\rm BE}$, for alternative supply voltages. From this, we can say that to get good stability with our given circuit we need an emitter voltage somewhere around 3V for Tr1, if we use potential divider biasing, as here.

If Trl is to operate with an I_c of $200\mu A$, we know that the emitter current will only be a little less than this. We shall ignore the effect of the relatively small base current for this calculation. (This is no trick of convenience, as one accusing reader expressed it! Simply that if we wasted our time calculating to the minutest figure, the resultant resistor values would not be available. The variations due to these current differences come within the preferred value 'spread' of resistors we propose to use.)

So, if you will grant us this much dispensation, ignoring $I_{\rm bl}$, we will calculate $R_{\rm el}$ as

$$\frac{V_{c1}}{I_c} = \frac{3V}{200\mu A} = \frac{3 \times 10^6}{200} = 15k\Omega$$

a preferred value of resistor.

Voltage gain

The d.c. voltage gain, $A_{\rm v}$, of Tr1, ignoring the effect of re, the internal emitter resistance (if Mr. Good will allow us to do so!), can be calculated from the collector load resistance divided by the unbypassed emitter resistance. That is,

$$A_{v} = \frac{18,000}{15,000} = 1.2$$

From a knowledge of the d.c. current gain of Tr1, which is operating at a collector current of $200\mu A$, we can calculate the true base current of the transistor. $h_{\rm FE}$ of a BC109 at a collector current of $200\mu A$ is typically 280.

$$I_{b1} = \frac{I_{c1}}{h_{EE1}} = \frac{200\mu A}{280}$$

which works out to approximately 0.7μ A.

The emitter voltage of Trl has already been

worked out: the base-emitter voltage will be 0.6V, as explained before; we shall aim at a current through the potential divider network, R1, R2, of five times the base current of Tr1.

Calculating R1, the lower resistor, this equals:

$$\frac{V_{b1}}{5 \times I_{b1}} = \frac{3.6v}{3.5\mu A} = \frac{36 \times 10^6}{35} = 1.03M\Omega$$

And again, the nearest preferred value will be chosen, $1M\Omega$.

The current flowing through R2 is the base current of Trl as well as this base current multiplied by five, which we decided as our desirable current through the potential divider. In other words, $6 \times I_b$.

So, R2 =
$$\frac{V_{cc}-V_{b1}}{6 \times I_{b1}}$$
 = $\frac{8-3.6V}{4.2\mu A}$

again working out approximately to $1M\Omega$.

Input resistance

In this context, we refer to transistor input resistance, which, in the case of the input resistance to the base of Trl is the effective emitter resistance of the transistor plus the effect of the unbypassed resistor.

Calculating first the internal emitter resistance of Tr1, we have the formula:

$$re = \frac{25}{I_c (mA)} = \frac{25}{0.2} = 125\Omega$$

 $R_{\rm in}$, ignoring R1 and R2, is equal to the effective emitter resistance plus the internal emitter resistance, multiplied by the a.c. current gain, $(R_{\rm eff}+r_{\rm e})\times h_{\rm fe}$.

Let's backtrack. Referring to Fig. 53, we bypass $R_{\rm ei}$ with a whacking great capacitor. Now, the stage gain

is going to be determined by $\frac{R_c.}{re}$ We can't do much

about R_e without mucking up the d.c. conditions (to quote verbatim from Mike's notes). If we want to modify the stage gain, therefore, we have to find an alternative way of doing it, and the solution is as shown in the drawing, a resistor, R_f, in series with a large value of capacitor, C_e, both across the emitter resistor, R_{e1}.

Now we find that R_f cannot affect the d.c. conditions. The amount of feedback in the circuit will depend on the sum of the feedback resistor, R_f , and the internal emitter resistance of the transistor, r_e . We have already agreed that re is 125Ω . If we make R_f 82Ω , the total emitter resistance (as far as a.c. is concerned) will be $82+125=207\Omega$.

Our stage gain will be controlled by the formula:

$$Av = \frac{R_c}{R_f + re} = \frac{18,000}{207} = 87$$

Reverting to the point where we backtracked and putting in our value of feedback resistor for the effective emitter resistance, we get an input resistance of $R_{\rm in} = 207 \times 360$ (the h_{fe} from the curves for the BC109) = 74520.

Having got the input resistance of the transistor settled. we can calculate the input resistance of the stage, which, you may remember, is obtained from the following formula:

$$\frac{1}{R_{IN}} = \frac{1}{R_{in}} + \frac{1}{R1} + \frac{1}{R2} = \frac{1}{74.5k\Omega} + \frac{1}{1M\Omega} + \frac{1}{1M\Omega} = 65.2k\Omega$$

Gain control

We can see that the gain of the circuit is obtained by altering the feedback resistor but this also affects the input resistance. Work out a few alternatives for yourself. When you do so, one fact will strike you: that is the major controlling factor of R1 and R2. We have already seen that there are decided limits to the amount of juggling that can be done, so in the next article we shall dispense with this pair of resistors altogether and show how the bias for the base of the first transistor can be derived from a single resistor fed from another part of the circuit.

Results

The final circuit, built up on 0.15in. matrix Veroboard gave us a measured gain of 90 (calculated 87), the input resistance was higher than calculated, and the output resistance when the circuit was fed from a low impedance source was approximately 522Ω . The output, measured and viewed on the oscilloscope, was 300mV.

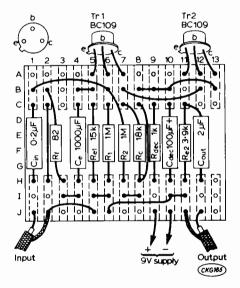


Fig. 54: Suggested layout of Fig. 53, on a 0.15in. matrix Veroboard.

Power supply details are as before, calculating R_{dec} to drop 1V from the sum of the various currents we have already considered, and decoupling with a large electrolytic to prevent the audio signals from modulating the supply. Coupling components are chosen from experience rather than calculated in exact detail. An input coupling capacitor could be $0 \cdot 2\mu F$ and for an output coupling capacitor, $2\mu F$ might be chosen.

TO BE CONTINUED

BACK NUMBERS

We regret that the back numbers department of P.W. has now closed and consequently we are unable to supply these. Requests for specific back issues can usually be included in our 'CQ' section; there is no charge for this but it is a service between readers and P.W. is unable to meet any of these requests. So many of the components you need for PW designs are in the new 1972 Electrovalue catalogue. Bigger, better than ever-Post free-10p.

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2N8704	NPN	Low power	100
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AC128	PNP	Low power	200
AC176	NPN	Low power	CTRA
AD149	PNP	High power	567
*AD161	NPN	Med. power	237
*AD162	PNP	Med. power	36p
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BC109	NPN	Low noise	121
BC168	NPN	Small signal	100
BC169	NPN	Low noise	110
BF194	NPN	RF amp.	140
BFY51	NPN	Med. current	20n
OA90	Ger, diode	RF detector	6n
OA91	11 11	General purpose	300
8D1		Silicon Rectifier 1 amp	90
WO2		Silicon bridge 1 amp	30p
	Ger = Germani	um. *Matched pair AD161/AD16	2 (60)
			1 200

417-470K E24

1Ω-10K E12 1Ω-10K E12

Values:
E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. E24: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 62, 72, 91 and their decades. Prices are in pence each for same ohmic value and power rating, NOT mixed values. (Ignore fractions of 1p on total value

E12 E24 E12 E24 E12

4:7:470K 4:7:10M 4:7:10M 4:7:10M 10:1M 0:22:3:9

Codes: C = carbon film high stability low noise

MO = metal oxide Electrosil TR5 ultra low noise WW = wire wound Plessey

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TYPE 3015F Seven segment indicator compatible with standard logic modules and power supplies. Figs. 0-9 from well illuminated filament segments to give character of 9mm height plus declimal point. Power requirement 8m Ar from 5v D.C. per segment. A limited number of alphabetical symbols also available.

In 16 lead DIL case

Sultant.

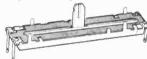
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0.1% to 0.01% tolerance. Prices and delivery details on request. Example—values up to 100K, between £1 and £2 nett.

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50p

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FLH201	(7401)	20p	FLJ141	(7474)	45p
FLH191	(7402)	20p	FLJ151	(7475)	45p
FLH291	(7403)	20p	FLJ131	(7476)	45p
FLH211	(7404)	25p	FLH221	(7480)	68p
FLH271	(7405)	25p	FLH231	(7482)	87p
FLH381	(7408)	25p	FLH241	(7483)	1.32
FLH391	(7409)	25p	FLH341	(7486)	33p
FLHIII	(7410)	20p	FLJ161	(7490)	80p
FLH351	(7413)	85p	FLJ221	(7491	
FLH121	(7420)	20p		AN)	1.28
FLH131	(7430)	20p	FLJ171	(7492)	85p
FLH141	(7440)	24p	FLJ181	(7493)	80p
FLL101	(7414)	1.22	FLJ231	(7494)	1.13
FLH281	(7442)	1 16	FLJ19!	(7495)	87p
FLH361	(7443)	1.45	FLJ261	(7496)	1.48
FLH371	(7444)	1.45	FLJ301	(74100)	1.64
FLH 151	(7450)	20p	FLJ281	(74104)	43p
FLH161	(7451)	20p	FLJ271	(74107)	52p
FLH171	(7453)	20p	FLK101	(74121)	48p
FLH181	(7454)	20p	FLJ201	(74190)	1.80
FLY101	(7466)	20p	FLJ211	(74191)	1.80
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5% 10% 5% 5% 2% 10% +1/20 Ω

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0-01; 0-012; 0-015; 0-022; 0-027; 0-053; 0-047;
0-056 each 8p.
0-068; 0-082; 0-1: 0-12; 0-15 each 4p.
0-18; 0-22; each 5p.
0-27; 0-33: 6p: 0-39 7p: 0-47 8p: 0-56 10p; 0-68 11p: 1µF13p.

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AF116 150	BC212L	USD.	MP8111	82p			2N3707	10p		(8p
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0.022μF, 0.033μF, 39, 0.04μF, 0.063μF, 0.1μF, 39, 0.1μF, 39, 0.22μF, 74μ. 0.05μF, 11μ. 0.01μF, 0.02μF, 0.022μF, 0.033μF, 0.047μF, 0.068μF, 39, 0.1μF 3 p, 0.15μF, 4 p, 0.22μF, 5p, 0.33μF, 6p, 0.47μF, 7 p, 0.68μF, 11μ. 1.0μF, 13p.

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0.29 0.29 0.37 0.37 0.47 0.43 0·15 0·15 0·15 0·15 SN7400 -SN7470 =8N7400 =8N7401 =8N7402 =8N7403 = 8N7470 = 8N7472 = 8N7473 = 8N7474 = 8N7475 0 15 0 15 0 18 0 18 0 18 0 18 = SN7404 =8N740 = 8N7403 = 8N7407 = 8N7408 = 8N7409 = 8N7410 =8N7476 =8N7480 =8N7481 =8N7482 =8N7483 0 67 0 97 0 97 1 10 0.29 = SN7486 0.82

BARGAIN TTL'S

=8N7410 =8N7413 =8N7416 =8N7417 =8N7420 =8N7430 =8N7440 =8N7441 =8N7442 0 48 0 48 0 48 0 15 0 15 0 67 0 67 0 67 0 67 =8N7490 A =8N7491AN =8N7492 =8N7493 0·15 0·67 -SN7494 0 77 0 77 0 77 1 75 0 97 0 40 0 55 1 25 -8N74950.67 0.67 1.95 1.95 1.95 0.97 0.97 =8N7496 = 8N7442 = 8N7443 = 8N7444 = 8N7445 = 8N7446 = 8N7447 = 8N7448 = 8N7496 = 8N74100 = 8N74104 = 8N74105 = 8N74107 = 8N74110 -8N74111 =8N74111 =8N74118 =8N74119 =8N74121 =8N74141 35 SN7450 0.15 -SN7451 0.15 = 8N7453 0.15 SN7454 0.15 = 8N74145 0.15 = 8N74150BN7460

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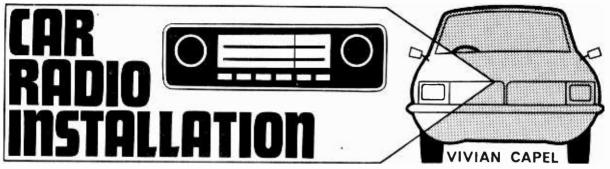


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PART 2

The main source of engine interference is, as we have said before, the distributor. Suppression is achieved by inserting resistance in the central high-tension lead from the coil. Modern British cars already have this fitted, often it takes the form of a resistive carbon brush for the rotor contact. No additional suppression should be fitted. In the case of older cars, and some of foreign manufacture, a suppressor should be fitted as a routine job when installing the car radio. A value of $10 \mathrm{k}\Omega$ is suitable which can be of the screw-in or cut-lead type, whichever is the most convenient.

The coil

The next point to consider is the ignition coil. This can generate interference which can be suppressed by fitting a $2\mu F$ capacitor. This should be of the metal-cased type, the case being secured to the car bodywork as near to the coil as possible. The capacitor lead, which must be kept as short as possible, must be connected to the low-tension lead coming from the ignition switch.

One type of capacitor-suppressor which is specially suited for this application, is the co-axial type. This is similar to the lead-through capacitors often found in television tuner-units. The metal case is clamped to the car bodywork, or better still the metal case of the coil, one connection is taken to the ignition switch-lead and the other to the coil. Thus the low-tension supply to the coil passes through the capacitor casing which thereby forms an effective barrier to interference pulses. The two capacitor leads must be kept as far apart as possible to prevent any inductive coupling to bypass the capacitor.

Dynamo suppression

Next, comes the dynamo. Again, interference can arise from this component, and a metal-cased ordinary or co-axial capacitor suppressor of $0\cdot 5\mu F$ should be used. What has been said about earthing the case and keeping leads short applies here too. Two leads come from the dynamo, one is the field-coil connection and the other is the output lead. On no account should the suppressor be connected to the field terminal as actual damage could result. One method of identification as to which is which, is that the field coils carry much less current than the output lead from the armature and so is wired with much thinner wire. The suppressor then, must be wired to the thick-wire connection.

We have described the three suppression points which should be attended to as a matter of routine on every car-radio installation. In most cases it will be found that interference is down to a very low level and the installation can be considered successfully completed. There are though the stubborn cases, and we will now see what can be done to silence these.

Stubborn cases

It may be that you are fitting a radio in a secondhand car that previously had a radio which was removed by the previous owner. Do not assume that the three points we have described were properly suppressed, check each one. Some garage mechanics have a habit of disconnecting or removing suppressors when looking for faults. In most cases the aerial will have been left on the car and of course this will be used in the new installation. As the majority of faults are attributable to the aerial, always check it in case of persistent interference. Check that the circuit from the aerial-rod to the radio-plug is continuous with a meter or other type of circuit tester. Check that the screening goes all the way and that there is no break or badly made join. Check that there is no leakage between the aerial and screen, and that the screen is a dead short to the bodywork.

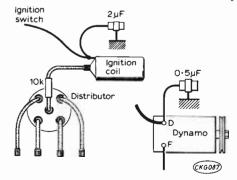


Fig. 2: Three routine suppression points for all radio installations.

Check that the part of the body that the aerial is mounted on and therefore the part that the screen is earthed to, is in fact in good electrical contact with the main bodywork. Sometimes new wings are fitted that are primed with undercoat before being bolted into place, and therefore make poor electrical contact. Lastly, check that all sections of a telescopic

aerial are in good contact with each other. Oxidization sometimes prevents this and the bottom section is 'on its own' with poor, if any, contact with those above it, thus resulting in a short aerial length, poor signal pickup and degraded signal/noise ratio.

If interference persists we must look elsewhere for the trouble. Cases have been known where after the most careful and extensive suppression the trouble has remained, and the cause has been traced to a burnt and pitted rotor-arm. It is worth replacing it if in doubt. Similarly the high-tension terminals in the rotor head may be worth checking and cleaning up.

If all is in order here, the plug leads may be radiating. A set of $5k\Omega$ resistive suppressors can be fitted, one in each plug lead, as near to the distributor as possible. It may even need a second set wired close to the plugs. Some makers use resistance wire for the plug leads and if this is the case (as can be checked with the meter) no further resistance will be of much help. Actually these resistive leads are better than wired-in suppressors, because the resistance is distributed along the complete length. Special h.f. filters may prove the answer.

It is assumed that the plugs themselves are in good condition, and of course these should be replaced at the stipulated intervals in the interests of petrol economy and engine performance. Old and worn plugs can cause interference, so if yours are due for a change, try a new set before getting too involved in suppression methods.

Other interference

Any interference that yet remains will very likely be due to causes other than the high-tension ignition circuit, if all the measures here described have been taken.

This can be ascertained by revving up the engine and then switching off the ignition. Any interference arising from the ignition circuits whether low or hightension will immediately cease. Any noise that continues as the engine slows down and stops must therefore be due to the dynamo or voltage-regulator. If it is still the ignition circuit that is giving the trouble, about the only thing left is inductive coupling between the l.t. ignition circuit and radio leads. One possibility that has been known, is coupling between the car wiring to the ignition-switch and the radio loudspeaker leads. Moving the latter may improve things, but in stubborn cases it sometimes pays to rewire the speaker with screened-cable. Some radios already have screened loudspeaker wire fitted by the makers, but often the screening is earthed to some point inside the metal case of the radio. Thus if any r.f. interference pulses are picked up by the screen they will be conducted inside the radio, and possibly affect the r.f. circuits. A worthwhile move is to disconnect the internal earth and to earth the screening to some external point on the case.

If in fact, the engine rev test eliminates the ignition circuits, we can absolve the dynamo if it is properly suppressed as we have described. This leaves the voltage regulator. It is usually easy to tell if interference is coming from this source because the sound has quite a distinctive character. It has what can best be described as a chattering sound, and it seems to affect the long-waves more than the medium waves.

Suppression will usually be effected by the connec-

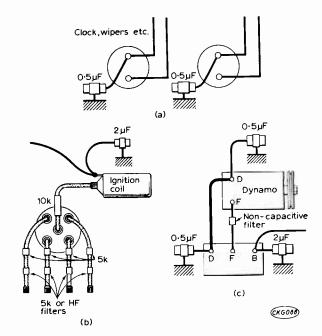


Fig. 3: All possible suppression points. Rarely necessary except in stubborn cases or FM radios.

tion of a 2μ F capacitor to the regulator terminal that goes to the battery. Do not connect it to the one that goes to the dynamo field coil.

If the main engine interference has been suppressed, as indeed it should be after the measures we have described, there may be a residual noise, especially with installations that have the aerial mounted at the front of the car. We have seen that this is more probable than those with rear mounted aerials, but in some cases the level may be higher than it should be. One possible reason for this is that the bonnet is not earthed electrically to the rest of the car. Thus its shielding effect is poor and the aerial is exposed to direct radiation from the engine. A couple of jumpers of copper braiding should be connected across the hinges to provide a low-resistance earth-path.

There are other causes of radio interference besides the engine and its necessary subsidiaries. A common one is the ticking of the electric-clock and the whine of windscreen-wipers. Fortunately, the cause is self-evident, and all that needs to be done is to fit a suppressor to the offending item. A metal-cased 0.5μ F should suffice as long as it is fitted as close to the accessory as possible with the shortest leads to it and to earth.

There is one final source of interference which has nothing to do with the car's electrical system, and this is static. It is easily identified because it is not present when the car is stationary, even though the engine is running, and it is present when the car is coasting with the engine switched off.

Although thus easy to recognise, it is not so easy to find the precise cause and effect a cure. It is generated by some part that is in poor electrical contact with the rest of the bodywork, and the problem is to find out which one. A common cause is the tyres; friction builds up an appreciable charge which then leaks through various paths producing the interference.

One possible cure, is the dangling-chain which a

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436

few years ago was widely used to prevent travel sickness. It discharges to ground any static that may be built up by the tyres or by the movement of the car body. This can be quite effective in the latter case but not so much in the former because of the rather long path from tyres to the chain. The most effective cure for tyre static is to provide a conductive path from the tyre tread to the car body. Painting the inside wall of the tyres with a conductive paint such as lead paint, or a rubber conductive paint, should do the trick. It is not necessary to paint the whole of the wall, several wide strips will do, providing they are taken from the wheel rims to the tread.

It may be that the wheels themselves are not in good contact with the body due to the wheel bearings. Packing them with graphite grease should make them conductive. Alternatively, special springs that are made for the purpose can be fitted inside the hubcaps. These form a bridge between hub-cap and axle.

In the case of the rear wheels it could be found that the back axle is not in electrical contact with the body, and if this is so it will be necessary to fit a copper-braid jumper from axle to body; the top bolts of the differential housing should prove a convenient anchoring point.

Other parts of the body or engine components can cause static through poor earthing, although it is usually the larger parts. Exhaust systems are a possibility and also the steering column.

From all this it may well be assumed that installing a car radio and suppressing all interference is a major job and one that is not to be lightly undertaken. This is not actually so, rarely will all the measures here described be found necessary, in the majority of cases the three routine suppressors we have before described will be sufficient. However, the stubborn ones do sometimes crop up, and it is hoped that in such an eventuality these tips will prove helpful.

It should be noted though, that f.m. car radios are more troublesome in the matter of car interference than a.m. All the suppression measures here described, except for static, should be carried out as routine when installing one. In addition to these an extra $0.5\mu F$ suppressor should be connected to the voltage-regulator, this time to the D terminal, and also a special non-capacitive suppressor to the F or DF terminals. As stated before, an ordinary capacitive suppressor must not be connected to the field-coil circuit.

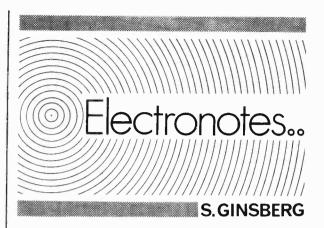
The resistive suppressors used for plug suppression will probably not be sufficient, and special h.f. filters may have to be used instead

TUT! TUT!!

TOM BAND CONVERTER. August 1972
Figure 1. Pin 8 on L2 should be connected to chassis and not to the junction of C2/C3. The wiring in Fig. 2 should be amended accordingly.

TAKE 20. August 1972

VR) is given as $5k\Omega$ in the circuit and $10k\Omega$ in the components list. Either will work but the circuit is correct and in preference VR1 should be $5k\Omega$.



NE of the problems (in certain applications) of using a standard typewriter keyboard to put characters or letters into a computer, is that you are stuck with exactly those characters. Thus a capital letter A is always exactly the same shape, thickness, height etc. It will be the same no matter who presses the key. It would be very useful if one could write, just as with an "ordinary" pen, and have the resultant "shapes" stored. These could be called up onto a visual display unit (v.d.u.) at will and could be transmitted to remote v.d.u.'s at another location. No longer would we be restricted to characters and numbers, and we could, if we wished, send a circuit diagram. It would be possible to send a signature which could be checked and verified by, say, a bank or security personnel.

Well, it can be done. As far as the "sender" is concerned he/she merely writes on a special tablet with an electronic pen. The method has been developed by the Siemens laboratories in Munical Any handwritten marks may be displayed immediately on a large cathode ray tube or stored in the computer, or transmitted to another computer or v.d.u.

The special writing tablet is a sheet of piezoceramic material. Ultrasonic pulses at about 500Hz are applied to two edges of the plate which are at right angles to each other. The result is that a wave front is generated from each of the two edges. These fronts travel across the plate but remain parallel to their edge of origin.

The pen doesn't write by pressure. It acts as a capacitive probe which reacts with these voltage fronts travelling across the piezoceramic at constant speed. Electronic circuitry interprets the pen's position in relation to the x-y axis formed by the wave fronts. The position of the pen is converted into signals used for the v.d.u. and computer.

Lasers are back in the news again. RCA has built a laser beam image reproducer. It is reckoned to produce film copies of TV pictures with a sharpness improvement of some forty times. The system has a line resolution of something like 20,000 lines. This is a considerable improvement over earlier RCA systems.

tems which had resolutions of 6,000 lines.

The other laser story comes from the Mecca of electronic innovation—Bell Laboratories in the U.S. Here, miniature gas lasers have been made which are only two inches long and 0.02in. in diameter. Operating at 6328 Angstroms, the first successful model employed helium-neon discharge. Using glass capillary tubing with a 430µm bore, gains of 2.7dBm were obtained.

TAKE 2®

JULIAN ANDERSON

A series of simple transistor projects, each using less than twenty components and costing less than one pound to build.

THERE is nothing much wrong with crystal microphones that a decent preamp will not cure. They are not as good as decent moving coil or ribbon types but it needs a discerning listener to notice the difference with one proviso: that the mike is properly matched to a high impedance source. Crystal microphones have got a bad name mostly because they are more often than not fed directly into a simple transistor amplifier with an input impedance of perhaps $10k\Omega$ providing a very bad mismatch. To get the best out of these inexpensive mikes, they must be connected to a really high impedance—at least half a megohm.

Figure 1 shows the circuit which comprises two sections. Trl and the associated components give the high input impedance as they are connected as a common-collector (or emitter-follower) stage. The input impedance to such a stage is roughly equal to the emitter resistor multiplied by the gain of the transistor and so we shall find that the input impedance is about $1 \mathrm{M}\Omega$ which will provide a very decent match to the crystal. Since we are dealing in high impedances we only require a low value coupling capacitor to d.c. block the input, in the circuit this is Cl, a $0.1\mu\mathrm{F}$ capacitor. The base bias for this stage is provided by R1.

The output of this stage will show no voltage amplification, in fact as far as voltage is concerned the output will be marginally less than the input. The transistor has used up its gain in converting the impedance, not in amplifying the signal in the conventional sense. The output from a crystal mike will vary with the type but for our purposes we shall take it as being 1mV. So we shall still be getting only a small output across the emitter resistor R2. To bring this up to a usable level to feed an amplifier we have to amplify the signal further. C2 d.c. blocks the output from Trl and this feeds to a series resistor. The function of this resistor is fairly complex but if it is omitted the signal will become distorted, the value is not critical but it will be found to lie around the level of that shown, $3 \cdot 3k\Omega$.

This resistor connects directly to the base of the conventional common-emitter amplifier Tr2 which, like Tr1, is a BC109. If this has a gain of 300 the output will be about 300mV, enough to drive pretty well any transistor amplifier. R4 provides the base bias and R5 acts as the collector load and the output is taken from the collector via the capacitor C3. C4 decouples the supply line. The version shown is for a negative earth type circuit but, if it is needed for a positive earth amplifier, all that is necessary is to change the transistors to 2N3702 types and reverse the polarity of the electrolytics. The gain may be slightly less but it should still be more than adequate.

As the input is a high impedance it is important that the input is screened. Crystal microphones are usually provided with a screened wire lead, the

No. 40 CRYSTAL MIKE PREAMP

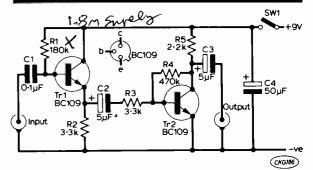
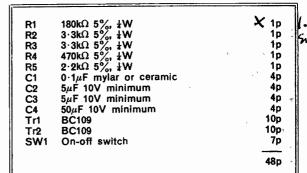


Fig. 1: The circuit of the crystal mike preamp.

★ components list



Prices are those recently advertised and may have changed. No allowance is made for minimum order, costs or for postage and packing and these points should be checked carefully before ordering.

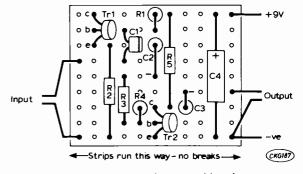
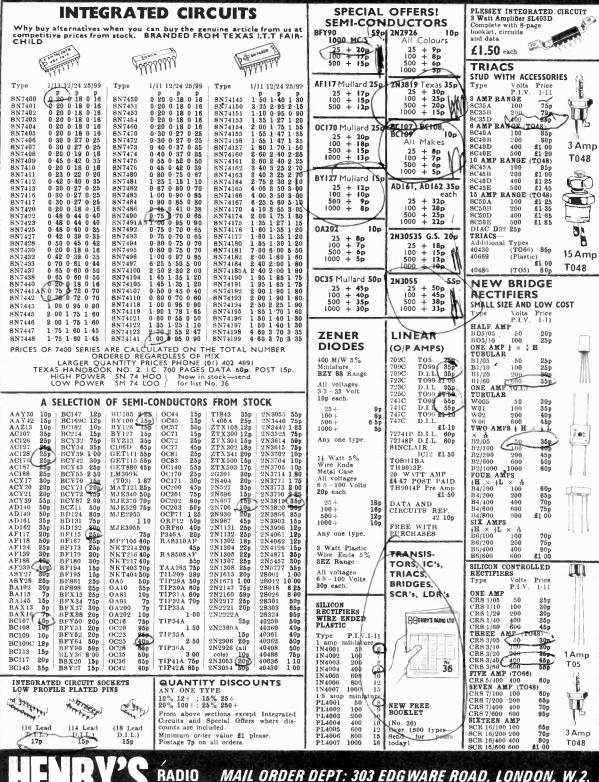


Fig. 2: A suggested component layout.

screening itself should go to the negative line (or positive line if that version is built). The output impedance is about $10k\Omega$ and although this is much lower than the input, a screened wire should still be used between it and the amplifier if it is more than a few inches long.

A suggested layout on a small piece of Veroboard is shown in Fig. 2 but building the circuit up on tagboard is just as good.

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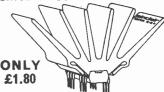
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MONTHLY NEWS FOR DX LISTENERS

NE of the many problems faced by the newcomer to the hobby of Shortwave Listening is the bewildering number of abbreviations which are used by more experienced listeners. Contrary to rumour these abbreviations are not used solely to confuse the uninitiated. Every branch of science and technology has its own set of abbreviations and jargon and Shortwave Listening is no exception to this.

To assist the uninitiated I will explain the meanings of some of the more common abbreviations which may be used on this page from time to time.

anncr Announcer. Ant Antenna, aerial. BC Broadcast or Broadcasting. Conditions (propagation). condx DX Long distance reception: **GMT** Greenwich Mean Time. hrd Hz Hertz, measurement of frequency equivalent to one cycle per second. ID, Ident Identification. International Reply Coupon, sent to help a IRC station with return postage. kHz 1,000Hz. -Latin American. LAMHz 1,000,000Hz. -Music. -Mx NA North America. Nx Card sent by a station verifying reception. OSL rcvd Received. Receiver. Rx SA South America. Sce Service (Home Sce., Overseas Sce., etc.). Sched Schedule. Sign off. s/off Sign on. s/on Shortwave. SW SWL Shortwave Listener. -TNX Thanks. TxTransmitter. Weather report. Wrp Wx Weather. Xmsn Transmission.

Adrian R. Pell has compiled the schedule of Radio Bangladesh which reads as follows:—

Xtal

Crystal.

Parallel transmission.

Address: Broadcasting House, 20 Green Road,

THE BROADCAST BANDS

Malcolm Connah

Frequencies in kHz • Times in GMT

Stations:

kHz.	m.	kHz.	m.
9690	30.96	15335	19.56
11620	25.82	15520	19.32
11650	25.75	17925	16.74

Programmes in English (Overseas Service):

0230 - 0300 GMT on 9690 and 15520 kHz. 1230 - 1300 GMT on 11620 and 17925 kHz.

1715 - 1800 GMT on 11650 kHz.

Listeners' Letterbox: Saturday at 1230, Sunday at 1715, Wednesday at 1230 and 1715.

Adrian also sent in the following log using his VEF 204 receiver:

7210 ICRC, Geneva via SBC at 1705.

9745 R. Baghdad, Iraq, s/on at 1930.

11650 R. Bangladesh at 1740.

11930 VOA, Okinawa, s/on at 2200.

11955 FEBA, Seychelles at 1730.

15335 R. Bangladesh in Bengali at 1430.

15520 R. Pakistan at 2050.

17705 VOA Monrovia, s/off at 0830.

Ian Howes of Lowestoft has used his R209 Mk. Il receiver and TV antenna to log the following stations:

3265 R.TV Congolaise, local Mx. at 2015.

3380 Malawi BC, local language at 2020.

4800 R. Lara, Venezuela, Spanish at 2345.

4880 Kinshasa, Congo local Mx. at 2315.

4940 Abidjan, Ivory Coast, local Mx. at 2320.

15245 R. Australia, ID in English at 2100.

15300 HCJB, Quito, Ecuador at 2030.

15345 TWR, Bonaire in Spanish at 2328.

R. Cooper of Foulden, near Thetford used his Philips 6 valve domestic recevier and a Bush 4 valve domestic to log the following stations:

11750 HCJB, Quito, Ecuador, English at 0830.

11915 BBC, Cyprus relay in English at 1645.

15020 R. Hanoi with Nx. in French at 1830.

15620 WINB, Red Lion at 2120.

15795 R. Nacional, Brazil, Mx. at 2010.

21500 RSA, South Africa, Nx. in English at 1000.

Richard Witney of Braintree is only 14 years old and is to be congratulated for sending in the following log:

6185 Radio Norway, Oslo at 1825.

7135 TWR, Monte Carlo at 1930.

9625 Radio Sweden in English at 1100.

9915 AIR, Delhi in English at 2100.

11725 Radio Bucharest, Rumania at 1515.

11910 Bangkok, Thailand at 1150.

Reports should arrive by the 15th of the month and be addressed to me at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.

DN THE



OT a very good month for listening between 1.8 and 30MHz, especially up the h.f. end. It seems that the summer evenings haven't lost their knack of making 20 metres the main dx band and logs indicate that most s.w.l's. appreciated this

David Knott (Middlesborough) enquires about DD3FK. Anyone know if DD works wonders in any other country?

Hands up if you live anywhere near Newtown in Montgomeryshire. Now hands down and get a pen! Seems that there's a newly-formed Amateur Radio Club which welcomes prospective members. All queries to: R. Litten, 512 Maesyrhandir, Newtown, Montgomeryshire.

Hot tip from Ian Hotchkiss (Hatfield) that there is an interesting net on Sundays on about 14300kHz (his dial not very accurate) at 1600 hrs. How's about

a quick ORX next Sunday?

Sam Elsdon (Halesowen) has been doing strange things with a 500 µA meter, diode and l.s. output terminals. (Never mind, the wounds will soon heal). Sam describes his antenna as "310ft. wire arranged in a sort of tapered square spiral in the loft". He asks, "Is there a name for that?" Well is there? Anyone care to suggest something? Gear, besides the above mentioned "thing" in the loft is, a CR70A and PR40. The 21MHz log for s.s.b. reads: CR3MD, CT1EAL, CT2AZ, F6VAA, FL8MM, JA1LCG, JW7FD, LU5AJ, OD5BA, PY2BC, PY7BF, TN8AU, VP8MM, W7GRH, YV4AGP, 4X4NX, 6W8AL, 9Q5DF.

Big round of applause for Stephen Worral (Stafford). He dared to listen on four metres. Two receivers are in use (anyone tried ham stereo transmissions yet?)-a B44 and FR-DX-400, the latter covering all amateur bands from 160 to 2 metres inclusive. (Cor! Think of winding the front-end coils by hand). Antenna for 70MHz is a four-over-four in the loft. Signals on 4 metres heard from G2FOS, G3LR, G3OJ, G3ANH, G3AOO, G3CDM/P, G3FBW/P G3JFO, G3JHM/A, G3JUB, G3JXN, G3KSU/P, G3KTH/P, G3LUP/P, G3MWQ, G3NAS, G3NEO, G3OCC, G30HC, G3OHH, G3OQT, G3OWW, G3PXP, G3RCQ/P, G3TQF, G3TVW, G3UGN, G3VPF/P, G3VPS/P, G3WCS, G3WXI, G3YBY, G4AGO, G4ALE, G4AYU, G5DF, GI3GLT, GW3ABR/P, GW3ITZ/P, GW3NWR/P, GW3OXD/P, GW3UCB/P, GW3XFY/M So who says there's nothing on 4 metres?

Up (or should that be down?) on 14MHz, Steve's log reads: CP1KRT, HP9AHD, JA1AEA, JA1AGH, JA3BQF, JA6LAE, KP9AID, KR6DO, 44 VK stations (Wow!), VQ9R, XT2AF, ZL1AFO, ZL1BER, ZL1BHB, all s.s.b. with the same gear except antenna; 75ft. end

fed at 20ft.

Alan Smith (Nelson) says he will continue to burn the midnight oil. Also committed to this awful pledge is a 680X and a 100ft. length of wire in the loft. Late night oil burnings to date raised 20 metre smoke signals from: CN8HD, CR4BS, CT1QN, EA3JE, EA6BH, EL2CB, EP2TW, HK4BNC, HK5AZA, HP9JW, IT9SPI, JA6GBB, JX2HK, JY9VO, KZ5SD,

THE AMATEUR BANDS David Gibson, G3JDG

Frequencies in kHz Times in GMT

PY8ZAA, PZ1DR, VE1YW, VK2AVA, VK3BCM, VK4SD, VP1BH, YS1MAX, YV5AK, ZD3M, 4X4VB, 5B4CDN, 5W1AU, 6Y5SR, 9K2AL, 9X5VA, 9Y4T. A quick singe on 15 metres brought in: AP2DU, EA7GF, JA6WAS, JH1DBU, JH1LZW, KV4FY, K9QFR, OD5CS, SV0WZ, SV0WP, VQ9R, VQ9MC, W9MEL, ZD8FM, ZB2A, ZD3M, 4X4BL, 9K2AL, 9X5VA all s.s.b.

Roger Hunt lives in a road I can't pronounce in Carmarthen. Listening utensil is a CR70A plus PR30. Antenna is a 190ft. long wire, end fed. Happenings on 28MHz include: CR60Z, PY2DVM, ZS6UR, 9J2DT, 9Q5SF. Down on the dreaded forty, where brave receivers cringe, signals were logged from: PY7BBD, PY8YT, TR8VE, ZD8TS, 4U1ITU, 5Z4LW, 9G1DY. I wouldn't mind living in a road I couldn't pronounce

just to hear things like that on 7MHz.

A self-confessed agent of Charles Molloy has written in! (Gad, these Medium wave devils are everywhere!) Gorgeous gear includes a FR-DX-400 with an 80 metre inverted V (wot, no dx loop wound on broom handles?). Said agent has purged his soul by listening on 3·5MHz to: CP1DN, CP6FG, CR4BS, CX2AX, KP4AST, KP4AN, KV4FZ, OY7JO, OY9LV, PY1HA, PY2ERS, PY3ABH, PY5CDZ, PY6SL, PY7BFN, VE1WV, VE2AL, VE3PT, VO1CQ, ZD8CS, ZD8RR, ZL4KF, 4X4UF, 3A2EE, 5R8AZ. Even 7MHz produced: ZL4KF, VK2AVA, YN9MQ and 9Q5BG. On two metres using a 6-ele homebrew and the same receiver: G4ART, G8AVH/P, G4BBH, G4BEL, G8BXX, G8CGG, G8CKC, G8DYX, G8FAB, G3ZVC. Name of the convert is **W. Waldron** (Cwmbran). Welcome home, brother Bill.

Alan Harper (Manchester) has a VEF204 receiver (I thought it was a new prefix for Rockall). Being b.f.o.-less, he uses a second receiver to get a beat note. Careful adjustment gave readable s.s.b. signals from CR6EM, EA5KR, HC1JB, HP1GS, KV4AD, PY2FCD, W8GL, ZF1WE, 4X4HT, 5N2ABG, 8P6DV, 9H1CV.

Ian Leslie (Mold) says that there are probably better logs about than his. Well, how does the one you didn't send in compare? Gear is an R107 and an end fed of unspecified length which bagged: CO2FA, CT1ZE, EA5IW, EP2MJ, JA2PJC, JX2AK, KV4AV, KZ5NG, M1B, OD5FA, PY4BSX, PY8BX, PZ1AX, T12GI, VE0NEC/MM, VE1KG, VE2ACP, VE6BB, VK2SG, VK5SV, VK6FP, VK7TR, VK9KA, WA1NGK/P/TF, YN1FI, YN1RSJ, YV10J, ZD3D, ZD7CE, ZP5KA, ZD3HT, 4S7AB, 4U1ITU, 4X4HT, 5Z4GG.

August is one of the busiest months in amateur radio. Six mobile rallies and three contests with another two early in September. Contests are: August 5-6, WAE c.w.; 12, 4 metre contest (worth a listen); 20, two metre s.s.b. contest; September 2-3, v.h.f. NFD; 2-3, IARU v.h.f. event; 3, qualifying round of the DF contest at Rugby.

Logs, in alphabetical order please, to arrive by the 15th of the month to:

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8 transistors—2 diodes 4" x 3" x 1"

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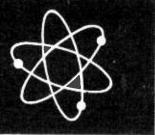
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EXPERIMENTERS CORNER



FULL WAVE POWER CONTROL CIRCUIT L. MacNAMARA

NUMBER of designs have been featured in past issues of Practical Wireless illustrating how s.c.r's (silicon controlled rectifiers) may be used to control such things as the speed of an electric motor and, although these devices worked perfectly well for their intended use, their versatility was limited when applied to light dimming control. In this particular application the s.c.r. has an obvious drawback in that it conducts only on one half cycle of the mains a.c. and the resulting 25Hz frequency appears as a flickering on the bulb, especially when run at low intensities. The obvious method to overcome this difficulty is to use a related device, the TRIAC, which conducts in both directions and the present article illustrates how a very simple and inexpensive power control unit may be assembled from a mere handful of components.

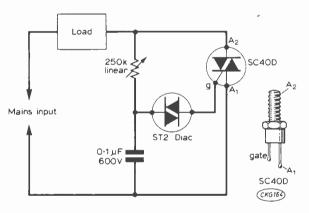


Fig. 1: Full wave power control circuit using a minimum of components.

At the heart of the unit is a G.E. D45D TRIAC (SC40D) which can effectively handle up to 6A at 240V. Basically it can be considered as consisting of two SCR's connected in inverse parallel either of

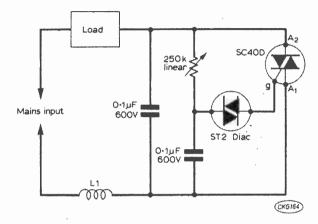


Fig. 2: The same circuit with r.f. suppression. L1 must be capable of handling 5A.

which can be triggered by a pulse applied at the gate. If the time at which this pulse is applied can be varied complete control of the power entering the TRIAC can be effected and this can be realised by a simple resistance/capacity variable phase shift network. An ST2 DIAC, which conducts when the voltage across it reaches about 20V of either polarity, acts as the trigger link between the TRIAC and the phase shift circuit. Fig. 1 shows how such a circuit can be designed using a minimum of components. However due to the sudden switching of the TRIAC from the "off" to the "on" state r.f. interference is produced and this can be a nuisance if the device is operated in the vicinity of radio or TV set. A simple inductance/capacity circuit was therefore incorporated to reduce the interference to a minimum and the prototype was constructed along the lines of Fig. 2. The complete unit was mounted in a plastic box with an ordinary 13A 3-pin socket attached. In addition a 5A fuse was incorporated in the plug to prevent the TRIAC being damaged should excessive current be drawn.

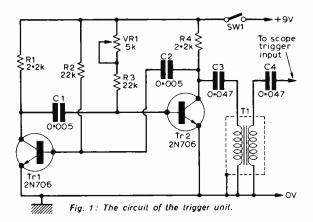
SCOPE TRIGGER UNIT

P. ROUSE

ANY government surplus oscilloscopes are of extremely high quality and well worth preserving, but sometimes they require an external trigger signal to produce a visible time base. The unit described below produces such a signal and also synchronises the oscilloscope input waveforms to the timebase frequency thus providing a very stable display for accurate measurements. It consists

of an astable multivibrator or square wave generator which is coupled to the trigger input on the oscilloscope by a small step-up pulse transformer. The circuit diagram is shown in Fig. 1 and its operation is briefly described.

The multivibrator produces square waves of either polarity at either collector but Tr2 is used because the frequency can be varied over a small range by



adjusting VR1. With the components specified, this range is from 5300Hz to 6500Hz which is quite adequate for synchronisation purposes. The waveform at this collector is shown in Fig. 2. The amplitude is only 8 volts and this must be increased to about 16 volts to trigger most of these old oscilloscopes. Furthermore, a fast pulse rise time is also required. In order to satisfy these two requirements a step-up differentiating pulse transformer must be used. This has the added advantage of producing both positive and negative voltage spikes thus allowing positive and negative triggering. The waveform between C4 and earth is shown in Fig. 3.

Differentiating pulse transformers should have very low leakage inductance and high permeability at their operating frequency. Consequently, a miniature ferrite i.f. transformer is ideally suitable but a Mullard ferrite pot core can be used. The i.f. transformer must be rewound in order to give the correct turns ratio and a diagram of the type used is shown in Fig. 4.

The i.f. transformer must be carefully dismantled and its windings removed. It is then rewound as follows. Firstly 100 turns of fine enamelled wire are wound onto the ferrite bobbin and the ends are soldered on to the lead-through pins in the plastic base. With thin enamel wire such as this, the heat of the soldering iron is sufficient to remove the enamel for easy tinning. Secondly 200 turns of the same gauge wire are wound on to the bobbin and the ends are soldered to the pins as before. The i.f. transformer is then reassembled.

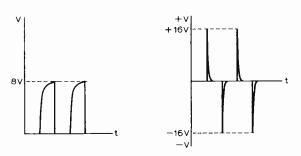
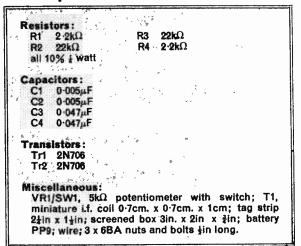


Fig. 2: (left) Waveform at the collector of Tr2. Fig. 3: (right) Waveform at C4.

* components list



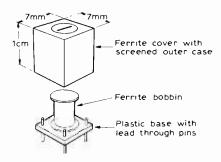


Fig. 4: Construction of the pulse transformer.

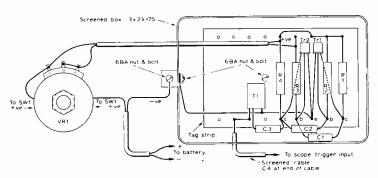


Fig. 5: A suggested component layout.

The circuit is wired up on a small piece of tag strip and mounted in an old tobacco tin for screening purposes. The switch on the back of the variable resistor VR1 is used as SW1 and the whole unit is small enough to be mounted inside the oscilloscope if required, in which case a hole of diameter ³8 in should be drilled in the front panel to hold VR1 and SW1. A 9 volt battery is used to provide power supplies and this voltage ensures that the transistors operate well within their limits thus providing good operational reliability. A diagram of the assembly is shown in Fig. 5.

GOING BACK... 1970 60 50 40 30 20 COLIN RICHES ARTHUR DOW

R W. W. Pickard from Lowestoft, Suffolk, tells us he still has an "Osram Music Magnet" with 2 volt valves and the old springy "Benjamin" bakelite valve holders, the set is in near perfect condition. He also has quite a lot of 2-volt valves, all sorts, given to him recently by a local dealer, and two old sets, one of which is a Gecophone, and the other he thinks a "Marconi". He says if any readers happen to be in Lowestoft at any time, they are welcome to see this ancient equipment and valves, if they will call at his business address, 192 London Road South, Lowestoft—an outfitters shop. Phone 5675 (closed Thursday).

He can go back to 1923, and had a 3-valve wireless set with plug-in coils then, and remembers well the first broadcast from 5XX Daventry. He remembers he had at that time an Amplion "Dragon" oak-horn loudspeaker—the price was 5 guineas in those days!

* * * *
Mr. S. D. Smith, 292 Whitton Avenue East, Greenford writes "The article "Going Back" prompts me to write down a few memories of my first home built wireless receiving sets. I used a similar device to the buzzer "thing" shown recently in "Going Back" to find a sensitive spot on the galena crystal. My buzzer was coupled inductively however and had no A or E terminals. Bypassing the galena, bornite Zincite etc., etc., we come to my first valve set. This was built on the bottom of a cut down wooden box from the grocers. A, E and phone terminals were P.O. type with steel wood screws for fixing. The valve holders were bought as separate sockets which were screwed separately into the box. The coils were 45 and 30 turns hank wound round the first three fingers and tied with thread. The aerial coil was clamped between two strips of wood by a centre screw which fixed it to the box. The reaction coil was tied to the end of a lath about 15in. long. The pivot screw (with washers) passing through a cotton reel fixed the assembly to the box above the aerial coil. A similar piece of wood was attached to the condenser, thus providing both slow motion and anticapacity control. One needed to be quite a good oarsman to avoid catching a crab! The condenser was straight line capacity and bunched the stations at the bottom end.

The phones were Sterling from the first World War for which I paid 2/6d. The valve I remember best had a horizontal assembly. It started to go blue above about 20V h.t. and was at that time treasured and only brought out for special occasions! My only

measuring instrument was an old Post Office galvo and a few resistances (resistors had not been invented). This was "calibrated" against new h.t. and g.b. batteries. All readings were strictly comparative. This instrument was almost indestructible and could be reversed on overload without ill effect. Once it was connected momentarily to the a.c. mains when on the 120V d.c. range. The only result was a slight demagnetisation, easily corrected by moving the paper "calibrations". The original variable grid leak consisted of a short length of glass tube, two corks, a bit of brass wire and burnt sugar. The degree of burning determined the resistance range—more or less. As I had no megohm meter I had to "suck it and see"!

I heard KDKA New York on two successive mornings at about 1.30 a.m. (My mother gave me a tongue lashing for my bleary eyes and threatened to smash up the gear.) It was many years before I heard America again and never since on medium wave. This experience made me very cynical about the "low loss" components which proliferated later and the degradation of the once honourable term High Fidelity. This was commented upon in the Marconi Book of Wireless (1936).

Memory ranges over unit sets, the Popular Wireless Combination Set which, after various mods. finished up with more SP and DP knife switches than a power house. It was more effective as an exercise in ingenuity and oneupmanship than a wireless set, but fun for all that. The Transatlantic Five with plug in McMichael r.f. transformers which never crossed the Atlantic for me, but which did give me a little ephemoral notoriety when I broadcast to all who cared to listen the results of the first election won by Labour. An early three gang neutrodyne I never did succeed in aligning and reverted to a separate aerial condenser, the old Transatlantic Five arrangement.

Later things became much more commercial and stereotyped so I will conclude with a two valve circuit which gave me a lot of fun for several years. In its final form it had a Ferranti choke and transformer and a BTH moving coil speaker, the latter replacing the Amplion horn, simple moving iron and the Blue Spot balanced armature speakers which served in turn. No screening was used and with all those fields and triode valves you may be able to imagine the stabilising devices that were tried out! Just before it was scrapped I found that reaction was very smooth and the set quite sensitive

if r.f. instability was induced over the whole wave band and the reaction coil used back to front. I did not recognise the possibilities of N.F.B. and thought no more about it for years. Hilversum and Paris came in well on L.S., i.e. speech clearly readable in any part of a normally quiet room. Brussels was somewhat weaker but it was only about 100 watts at that time, if my memory is not at fault. It wasn't one of the giants anyway.

I wonder how it would work now? Pentodes, ferrite aerials screened coils and valves, reaction to the first anode and three tuned circuits. Perhaps I'll try it one day if I can find a parafeed 3-1 valve

transformer. The AF 5 was a bit bulky.

As an after thought-how many remember the very formal announcement --- "Hallo, Hallo, Hallo, this is 2LO calling; this is the London Station of the British Broadcasting Company calling. You will now hear the news, copyright by Reuter, Press Association. Exchange Telegraph and Central News"?

Marconi Kemp Stamp

On September 13th the Post Office will issue a set of four commemorative stamps. One of these, valued 712p, commemorates the 75th Anniversary of the first wireless transmissions across water. These tests were carried out by Marconi and Kemp from Lavernock Point, near Barry, to Flatholm Island in the Bristol Channel and Brean Down in Somerset.

The Barry College of Further Education Radio Society commemorates these tests annually by operating amateur radio stations from these historic locations. The Society was responsible for suggesting to the Post Office the issue of this stamp, and are justifiably proud of their association with its issue.

In connection with this issue, they are making available a special commemorative first day cover service from Flatholm Island. For the first time the



The design on the envelope in blue and red.

Post Office has granted a postmark from Flatholm Island. The special postmark will include their callsign GB3BCT (Bristol Channel Tests)—the callsign to be used by their station operating from the island on the 13th September. The first day cover is an embossed design showing the location of the three

This unique combination should be of great interest to radio enthusiasts and philatelists alike and the Society will provide the special commemorative envelope, the 712p Marconi stamp, address and post from Flatholm Island on the first day of issue for . . . 20p.

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Vintage Records

Once again we ask readers to let us know the titles and labels of any pre-1930 78's they may have as we have received a number of queries asking the whereabouts of such records.

In addition, if anyone has any gen or pictures of "talking machines" cylinder or disc, we would be glad to hear from them as we are considering writing a feature on this subject in a future "Going Back."

Vintage CA CA CA

EQUIPMENT FOR DISPOSAL

EQUIPMENT FOR DISPOSAL
....photographs and constructional details of the "All Concert de luxe" receiver by Percy W. Harris, about 1924 also two Scott-Taggart books. Offers to—A. E. Robinson, 34 Haddon Way, Carlyon Bay, St. Austell, Cornwall.
...3 volumes of Harmsworth Wireless Encyclopedia in good clean condition. Together with other items. See for details please—F. L. Harris, 80 Queens Walk, Ashford, Middlesex, TW15 3.B.
...Buzzer Wavemeter similar to the one in your column a short time ago, various wirewound, paneake and plug-in coils by igranic and others, Ferrant and Form Intervalve transformers and chokes, filament rheostats, ebonite panels etc. and several examples of single and multiple variable condensers by Ormond and Newdy. These later being mechanical curiosities in their own right—and mostly brand new. I think there are a couple of complete (?) receivers somewhere in the attics but have not yet come across them. Any offers—J. C. Priest, 21 Levens Grove, Blackpool, Lancs. FY1 3L A.
..." Wireless Telegraphy and Telephony; Charles R. Gibson; Seeley Service, 1914.
"The Armstrong Super Regenerator Circuit; George JELT2, Radio Directory and publishing Co. New York, 1922. "The Art and Science of the Gramophone" Gaydon 1928.—V. C. Clever, 11 Robin Dene, Manor Road, Brighton, BN2 5EX.

INFORMATION WANTED

"Marconiphone Receiver Type 22 Inst. No. S.C.2651". Unfortunately I have only one coll, longwave, 1000-2200 metres. This is a black ebonite cylinder enclosing coil and beautiful little reaction coil (adjustable from dial on front) marked "Aerial reaction unit, Marconiphone Pat. No. 217042". I would be very pleased to hear of a medium wave coil, 200-500 metres. It clips on to a base with four clips. The two valves fit into one holder mounted on sponge rubber.—W, W. Pickard, 333 Long Road, Lowestoft. Suffolk. Road, Lowestoft, Suffolk.

EQUIPMENT WANTED

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...bright emitter valves or dull emitter with exhaust pips on top, Also horn to fit diaphragm-type speaker, and pre-1920 components.—P. Beckley, 14 Beechdale Road, Newport, Monmouthshire, NPT 8AE.

"J. S. T."

In the article last month on John Scott-Taggart the call-sign of his amateur radio station shown on page 321 should have been LU (later LUX). The receiver illustrated on page 322 was the ST400, not the ST300.

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74 1.40 2N3391A 0.2 76 0.75 2N3392 0.1	2 2N5458 0-88 AF186	0.40 BCY30 0.85 BFV 0.85 BCY31 0.40 BFX	15 0.85 8N7410 0.20	0.18 8N7443 1.04 0.88 8N7483 1.80 1.0
74 0.75 2N3393 0.1 35 0.75 2N3394 0.1	8 3N128 0.68 AF239	0.86 BCY32 0.60 BFX 0.68 BCY32 0.50 BFX	29 0.25 SN7412 C-48	0.42 8N7445 1.85 1.70 8N7485 2.50 2.8
1 0.95 2N3402 0.1 6 1.00 2N3403 0.1	7 3N199 1.86 AF279	0.47 BCY33 0.80 BFX 0.47 BCY34 0.85 BFX	37 0-88 8N7416 0-45	0.40 SN7447 1.80 1.20 SN7488 12.40 11.8
84 0.98 2N3404 0.2 88A 0.40 2N3405 0.2	4 3N141 0.69 AFY42	0.65 BCY38 0.40 BFX 0.55 BCY39 0.80 BFX	63 2-48 BN7420 0-20	0.18 SN7449 Price to be SN7490 (0.80) 0.7
04 0.28 2N3414 0.1 04 0.28 2N3415 0.1	0 3N143 0.64 AL102	0.75 BCY40 0.50 BFX 0.70 BCY42 0.15 BFX	84 0-24 BN 7422 0-48	0.42 SN7450 0.20 0.18 SN7492 0.80 0.7
156 0.75 2N3416 0.1 156A 0.75 2N3417 0.2	5 3N153 0.74 ASY26	0.25 BCY43 0.15 BFX 0.30 BCY58 0.18 BFX	86 0.24 BN7425 0.48	0.48 8N7452 0.20 0.18 8N7494 0.85 0.7
7A 0.80 2N3570 1.2 1 3.25 2N3571 1.1	5 3N159 1.00 ASY28	0.25 BCY59 0.19 BFX 0.80 BCY66 0.58 BFX	88 0.25 8N7427 0.48	0.28 BN7453 0.20 0.18 BN7495 0.85 0.7 0.42 BN7454 0.20 0.18 BN7496 1.00 0.9
0.26 2N3572 0.9 0.34 2N3702 0.1	7 3N200 2:07 ASY50	0.20 BCY67 0.82 BFY 0.35 BCY70 0.17 BFY		JB-MIN ELECTROLYTIC
0·15 2N3703 0·1 0·15 2N3704 0·1	A8Z21	0.55 BCY71 (0.22 BFY	17 0.90 Axial lead	0:84/84: 1/40: 1.8/95: 0.5/38: 0.0/89: 4/39
0.25 2N3705 0.1 0.29 2N3706 0.9	BC108	0 19 BCY87 3 47 BFY 0 11 BCY88 2 40 BFY	20 0.50 25/25; 32/10; 32/	40; 40/16; 50/64; 50/25; 50/40; 64/10; 80/16; 80/98
6 0·10 2N3707 0·1	BC113	0-19 BCY89 0-80 BFY 0-13 BCZ10 0-35 BFY	37 0.20	125/10; 200/10.
8 0.13 2N3709 6.9 9 0.88 2N3710 0.9 1 0.80 2N3711 0.9	10050 0 E0 BC115	0.12 BCZ11 0.50 BFY 0.15 BD115 0.75 BFY	13 0.65 Prz.	SILICON RECTIFIERS 100 200 400 600 800 1000 1200
11 0.80 2N3711 0.9 18 0.21 2N3712 0.9 18A 0.80 2N3713 1.0	40201 0.08 BC116A		51 0·16 1A 8a	9p (10p 11p 12p 15p 20p
0 0-50 2N3714 1-1 1 0-55 2N3715 1-2	40310 0.87 BC118	0.21 BD123 0.82 VFY 0.11 BD124 0.60 BFY 0.27 BD130 0.50 BFY	53 0·15 6A	7 17p 20p 22ip 25p 27p 30p 35p
0.15 2N3716 1.8 0.17 2N3773 2.4	40362 0.40 BC121	0.20 BD131 0.75 BFY 0.25 BD132 0.75 BFY	34 0.41 15A 86p	45p 48p 55p 65p 75p 87p — 80p 90p \$1.00 \$1.40 \$1.70 \$2.75 —
6 0.17 2N3773 2.46 8 0.80 2N3774 8.8; 9 0.20 2N3775 4.1;	40389 0.87 BC125	0.15 BD135 0.88 BFY 0.20 BD136 0.44 BFY	6 0.22 lamp and 3 amp	are plastic encapsulation.
9 0-14 2N3776 5-9 0 0-14 2N3777 4-8	40394 0.58 BC132 40395 0.50 BC134	0.80 BD137 0.47 BFY 0.11 BD138 0.55 BFY	78 · 0.86	DIODES & RECTIFIERS
0 0-30 2N3778 2-2: 1 0-32 2N3779 3-1:	40407 0.81 BC135	0·11 BD139 0·62 BRY 0·15 BD140 0·73 BSX	9 0.88 IN34A 10p	AA119 7p BAX16 12ip F8T3/4 22is AA129 15p BAY18 17ip OA5 17p
1 0.29 2N3780 4.56 2 0.20 2N3781 8.65	40409 0.49 BC137	0.15 BDY10 1.25 B8X 0.24 BDY11 1.50 B8X	0 0.14 IN916 7p	AAZ13 12p BAY31 7p OA10 20p AAZ15 12p BAY38 95p OA9 10p
4 1.27 2N3782 3.8° 2 0.16 2N3789 1.7° 3 0.16 2N3790 1.9°	40411 1 40 BC141	0.80 BDY17 1.50 BSX 0.84 BDY18 1.75 BSX	7 0.84 IS113 15p	AAZ17 10p BY100 15p OA47 8p. BA100 15p BY103 0A70 7p
0·16 2N3790 1·9 0·20 2N3791 2·0	40468A (0.85) BC143	0.18 BDY19 1.97 BSX 0.21 BDY20 1.00 BSX	8 0.25 I8120 12p 9 0.47 I8121 14p	BA102 25p BY122 47ep OA73 10p BA110 25p BY124 15p OA79 7p
0.20 2N3792 2.20 0.22 2N3794 0.10 0.22 2N3819 0.22	10001 AFF BC144	0.24 BDY38 0.65 B8X 0.21 BDY60 0.90 B8X	0 0.68 IS130 Sp 9 0.78 IS131 10p	BA114 15p BY126 15p OA81 8p BA115 7p BY127 17p OA89 15p
0.22 2N3794 0.10 0.22 2N3819 0.25 0.25 2N3820 124	40603 0-49 BC148 40604 0-45 BC148	0·10 BDY61 0·85 BSX 0·09 BDY62 0·75 BSX	1 0.42 IB920 7p	BA141 17p BY164 57p OA90 78 BA142 17p BYX10 22p OA91 77
0.25 2N2923 0.69 0.90 2N3824 0.78	40636 0-88 BC153	0.11 BF115 (0.25 BSX 0.10 BF117 0.43 BSX	7 0.20 18923 12p	BA144 12p BYZ10 85p OA95 78 BA145 17p BYZ11 82p OA200 78
0.80 2N3826 0.28 0.20 2N3854 0.16 0.88 2N3854A 0.16	AC107 0-35 BC157 AC113 0-16 BC159	0·11 BF119 0·58 ASX 0·09 BF121 0·25 BSW 0·10 BF123 0·27 BSY	0 0.28	BA154 12p BYZ12 30p OA202 10p BAX13 5p BYZ13 25p TIV307 50p
0.86 2N3855 0.16 0.82 2N3855A 0.16	AC117 0-20 BC159	0·10 BF123 0·27 B8Y2 0·10 BF125 0·25 B8Y2 0·11 BF127 0·27 B8Y2	5 0.20 OPTOELEC	TRONICS SLIDE POTENTIOMETERS
1.00 2N3856 0.16 1.10 2N3856A 0.16	AC121 0-11 BC167B	0.11 BF152 0.20 BSY2 0.10 BF153 0.29 BSY2	7 0-15 MINITRON 3915	F 7-SEGMENT 58 mm, TRACK
2·15 2N3858 0·16 0·17 2N3858A 0·16	AC128 00:20 BC168C	0·10 BF154 0·16 BS13 0·11 BF158 0·28 BSY3	8 0.15 May be driven	PIN DIL) 22.00 by 8N7447 SINGLE GANGED, LOG or LIN 1k to 1M. 40p each. TWIN GANGED, LOG or LIN
0.84 2N3859 0.16 0.80 2N3859A 0.16	AC142K 0.25 BC170	0.11 BF159 0.27 BSY1 0.11 BF160 0.23 BSY1	1 0.25 GNP-7AH COL	D CATHODE 1k to 500k film each
0.70 2N3860 0.16 0.60 2N3866 0.70	AC152V 0-15 BC172	0·18 BF161 0·85 BSY5 0·11 BF163 0·20 BSY5	U WO DROTMAT. P	AN 75n MULLARD C280 M/FOIL
0.40 2N3877 0.25 0.40 2N3877A 0.26	AC153K 0.22 BC182L	0·10 BF166 0·85 B8Y5 0·10 BF167 0·18 B8Y6	0.15 TIL 200 TIGH	CAPACITORS
0.58 2N3878 1.22 A 0.61 2N3879 1.91 0.27 2N3900 0.20	AC176 0.16 BC183L	0.09 BF173 0.19 B8Y7 0.09 BF177 0.25 B8Y7	0.40 DIODE. Made by	TEXAS INST 0.068, 0.10
A 0.80 2N3900A 0.21	AC187K 0-17 BC184L	0 11 BF178 0 31 BSY7 0 11 BF179 0 38 BSY9	A 0.09 POOCE PROPER	9 0.47
0·37 2N3901 0·32 A 0·18 2N3903 0·22 A 0·18 2N3904 0·22	ACY18 0-15 BC187	0-25 BF180 0-35 BU10 0-25 BF181 0-34 BU10 0-11 BF182 0-40 C111	2.25	1μF 14p
0.20 2N3905 0.21 0.20 2N3906 0.22	ACY20 0.20 BC205	0·10 BF183 0·40 C407 0·11 BF184 0·17 C424		0.15 0.1 2.2µF 289
0.20 2N4036 0.40 0.20 2N4037 0.85	ACY22 0-18 BC207 ACY22 0-13 BC268	0·10 BF185 0·17 C425 0·09 BF194 0·14 C426	0.86 21 × 31 in	fatrix Matrix WIRE-WOUND RESISTORS 17p 28p 2.5 watt 5% (up to 270 ohms
0.22 2N4058 0.12 0.20 2N3059 0.90	ACY30 0.42 BC209 ACY30 0.42 BC211	0·10 BF195 0·15 C428 0·80 BF196 0·15 C444	0.25 31 × 31in	25p 25p only), 7p
0.82 2N4060 0.11 0.11 2N4061 0.11	ACY40 0.17 BC212K	0.10 BF197 0.15 D40N 0.12 BF198 0.18 GET1	34 × 5in 0-62 5 × 17in (Plain)	30p 29p 10 watt 5% (up to 26k Ω only),
0·12 2N4062 0·11 A 0·17 2N4302 0·25	ACY44 0.81 BC214L	0 14 BF199 0 18 GET1 0 16 BF200 0 40 GET1	Vero Cutter 45p	30) ZOP
0.45 2N4303 0.82 1.20 2N4913 0.80	AD140 0.50 BC237	0.08 BF224J 0.14 GLT1 0.08 BF225J 0.19 GET1	5 0.50 The insertion 100	ls (·1 and ·15 POTENTIOMETERS Carbon:
0·12 2N4914 0·87 2 0·12 2N4915 0·95	AD142 0.50 BC239 AD142 0.50 BC251	0.08 BF257 0.22 GET1 0.20 BF238 0.22 GET5	0 0.40	Log. or Lin., less switch, 16s.
0·17 2N4916 0·11 0·17 2N4917 0·17	AD149V 0.58 BC253	0.18 BF244 0.16 GET5 0.28 BF245 0.83 GET5	6 0-20 DISCHARGE	IGNITION Wire-wound Pots (3W), 1887 Twin Ganged Stereo Pots, Log.
4 0.18 2N4918 0.50 4A 0.29 2N4919 0.58	AD150 0.55 BC258	0.08 BF246 0.48 GET5 0.08 BF247 0.49 GET8	3 0.15 (As published i	in P.E. Nov. or Lin. 40p.
5 0.26 2N4920 0.60 5A 0.28 2N4921 0.50	AD161 0-36 BC259 AD161 0-36 BC261	0.11 BF254 0.14 GET8 0.20 BF255 0.15 GET8	0 0.35 71). Complete	50p. PRESETS (CARBON)
6 0.18 2N4922 0.55 6A 0.28 2N4923 0.60	AD162 \ (0.60 BC263	0.18 BF257 0.41 GETS	7 0.20	0.1 Watt 6p VERTICAL 0.2 Watt 6p OR
7 0.18 2N4926 0.90 2N4927 1.00	API06 0.94 BC.01	0.42 BF258 0.46 GET8 0.84 BF259 0.48 GET8	" A Carbon Pilm	0.3 Watt 74p HORIZONTAL
3 0.75 2N4929 2.28	AF109R 0-35 BC302 AF114 0-25 BC303	0.27 BF261 0.28 MA800	1 0.16 watt 5%, 1p.	W, 1W & 2W E12 Series P.W. Digital Clock
3 0-12 2N4930 2-25	AF115 0.25 BC304	0.43 BF270 0.25 MA800	4 0'4/ 1 watt 5%, 11n	1 1
transistors onl) 15p. ex	tra per pair. Prices subject	th (Air) 65p (MIN.) Matching to alteration without prior	3 0.47 watt 2%, M/O 4p charge 1 watt 10% 21p notice. 2 watt 10% 5p	W & W E24 Series. £20.50 April '72 issue. (inc. P. & P.)
01-452 0161/2/3		A. MARSHAL	L & SON	< CALLERS-WELCOME
Telex 21492	28 CRICKLEW	OOD BROAD	WAY, LONDO	N, N.W.2 HRS. 9-5.00 MONFRI. 9-5.0 SAT.

Sinclair Project 60



Active Filter Unit

(A.F.U.)

Built and tested post free **f.5.98**

The value of an efficient filtering system cannot be over emphasized in these days of very high quality reproduction since there are so often occasions where its use can mean the difference between comfortable and uncomfortable listening. On the low pass side the Sinclair A.F.U. will effectively reduce hiss from radio or tape, cut out heterodyne whistles on A.M. reception, greatly reduce record surface noise and other imperfections; on the high-pass side it will cut out motor rumble and other spurious low frequency intrusion. The unit is for use between pre-amp (including tape pre-amps) and power amplifiers, and operates in two sections, both stereo. The cut-off frequencies are continuously variable, and since attenuation in the rejection band is rapid (12dB/octave) there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is as easy to mount as the stereo 60 pre-amp/control unit which it matches in styling, along with the Stereo FM Tuner.

SPECIFICATIONS

The A.F.U. employs two Sallen and Key type active filter stages, one rumble (high pass) and one scratch (low pass). The two stages use complementary transistors to minimise distortion.

Supply voltage: 15 to 35 volts. Current

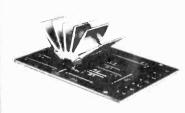
Gain at 1kHz: Filters flat 0.98 (—0.2 dB)

HF cut off: (—3dB) variable from 28 kHz
to 5 kHz at 12 dB/octave.

LF cut off: (—3dB) variable from 25Hz to 100Hz at 12dB/octave.

Distortion: at 1kHz (35 volt supply) 0.02% at rated output.

SuperIC.12 Integrated circuit high fidelity amplifier



Having introduced Integrated Circuits to hi-fi constructors with the IC.1.0, the first time an IC had ever been made available for such purposes, we have followed it with an even more efficient version, the Super IC.12, a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up, F.M. radio or small P.A. set up, etc. The free 40 page manual supplied, details many other applications which this remarkable IC. make possible. It is the equivalent of a 22 transitions and the supplications which the supplications of the equivalent of a 22 transitions.

sistor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC 12 is compatible with Project 60 modules which would be used with the Z.50 and Z.30 amplifiers. Compilete with free manual and printed circuit board.

SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak). 6–80. Frequency Response: 5Hz to 100KH2±1d8. Total Harmonic Distortion: Less than 1%. (Typical 0-1%) at all output powers and frequencies in the audio band (28V). Load Impedance: 250 Kohms nominal. Power Gain: 90dB (1,000,000,000 times) after feedback. Supply Voltage: 6 to 28V. Quiescent current: 8mA at 28V. Size: 22×45×28mm including pins and heat sink...

Manual available separately 15p post free,

With FREE printed circuit board and 40 page manual.

£2.98 Post free

Project 605



The easy way to buy and build Project 60

Project 605 is one pack containing one PZ5 two Z30's, one Stereo 60 and one Masterlink. This new module contains all the input sockets and output components needed together with all necessary leads cut to length and fitted with neat little clips to plug straight on to the modules. Thus all soldering and hunting for the odd part is eliminated. You will be able to add further Project 60 modules as they become available adapted to the Project 605 method of connecting.

Complete Project 605 pack with comprehensive manual, post free £29.95

Everything you need to assemble a superb 30 watt high fidelity stereo amplifier without having to solder.

Sinclalr Radionics Ltd, London Road, St. Ives, Huntingdonshire PE17 4HJ. Tel: St. Ives 64311



the world's most advanced high fidelity modules

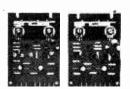
Z.30 & Z.50 power amplifiers

Built, tested and guaranteed with circuits and instructions manual 2.30 £4.48 2.50 £5.48

The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to provide unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at 15w (8 Ω) and all lower outputs. Whether you use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and are intended for use principally with other units in the Project 60 range. Their performance and design are such, however, that Z.50s and Z.30 may be used in a far wider range of applications.

SPECIFICATIONS (Z.50 units are interchangeable with Z.30s in all applications). — Power Outputs: Z.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts.
Z.50 40 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts.
Frequency response: 30 to 300,000Hz + 1dB. Distortion: 0.C2% into 8 ohms. Signal to noise ratio: better

than 70dB unweighted. Input sensitivity: 250mV into 100 Kohms (for 15w into 8Ω). For speakers from 3 to 15 ohms impedance. Size: 14 x 80 x 57mm



Stereo 60 Pre-amp/control unit

Designed specifically for use on Project 60 systems, the Stereo 60 is equally suitable for use with any high quality power amplifier. Since silicon epitaxial planar transistors are used throughout, a really high signal-to-noise ratio and excellent tracking between channels is achieved. Input selection is by means of press buttons, with accurate equalisation on all input channels. The Stereo 60 is particularly easy to mount.

SPECIFICATIONS—Input sensitivities: Radio – up to 3mV. Mag. p.u 3mV correct to RTA.A. curve £1dB 20 to 25,000 Hz. Ceramic p.u – up to 3mV. Aux – up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE+12 to —12dB at 10KHz BASS +12 to —12dB at 100Hz Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm

Built, tested and guaranteed.

Built and tested. Post free.

£25

£9.98

Project 60 Stereo F.M. Tuner

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 plus 20 m.l.C. Tuning range: 87 to 108MHz Sensitivity: 7µV for lock-in over full deviation. Squelch level: Typically 20µV. Signal to noise ratio: >65dB. Audio frequency response: 10Hz = 15KHz (±1dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: 2µV. Cross talk: 40dB. Output voltage: 2 x 150mV R.M.S. maximum Operating voltage: 25-30VDC. Indicators: Stereo on: tuning. Size: 93 x 40 x 207mm.



Power Supply Units

Designed specifically for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 or PZ.8 where a stabilised supply

System	The Units to use	together with	Units cost	
Simple battery record player	Z.30 Crystal P.U., 12V batter volume control, etc.		y £4.48	
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U. volume control, etc	£9.45	
12W RMS continuous sine wave stereo amp for average needs	2 x Z.30s, Stereo 60; PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90	
25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers	2 x Z.30s, Stereo 60; PZ.6	High quality ceramic or magnetic P.U . F.M. Tuner, Tape Deck, etc.	£26.90	
80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms)	2 x Z.50s, Stereo 60; PZ.8, mains transformer	As above	£34.88	
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43	







Guarantee

If, within 3 months of purchasing any product direct from Sinclair Radionics Ltd., you are dissatisfied with it, your money will be refunded at once. Many Sinclair appointed Stockists also offer this same guarantee in co-operation with Sinclair Radionics Ltd.

Each Project 60 module is tested before leaving our factory and is guaranteed to work perfectly. Should any defect arise in normal use, we will service it at once and without any charge to you, if it is returned within two years from the date of purchase. Outside this period of guarantee a small charge (typically £1.00) will be made. No charge is made for postage by surface mail. Air Mail is charged at cost.

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Please send	
Name	
Address	PW 972

PZ.5 30 volts unstabilised £4.98

£7.98

£7.98

PZ.6 35 volts stabilised

PZ.8 45 volts stabilised (less mains transformer)

PZ.8 mains transformer

TRANSISTORS, ZENER DIODES etc. **Fully guaranteed** TRANSISTORS, Z OA5 200 OC71 129 IN823A 41:80; OA70 109 OC72 209 IN4785 509; OA71 109 OC75 209 IN4785 509; OA73 109 OC76 209 IN4785 509; OA74 79 OC81 209 IZMT10 389; OA76 (A) 109 OC76 209 IZMT10 389; OA79 (A) 109 OC76 209 IZMT10 389; OA79 (A) 109 OC76 209 IZMT10 389; OA79 (A) 109 OC81 209 IZMT10 389; OA91 79 OC81 209 IZMT10 389; OA91 79 OC81 209 IZMT10 389; OA200 209 OC82 259; OX188 379; OA202 109 OC83 259; OX188 379; OA201 209 OC83 259; OX180 259; OA210 259 OC84 259; OX1874 649; OA211 309 OC162 509; OX1874 649; OA210 259 OC84 259; OX1874 649; OA210 259 OC84 259; OX1874 649; OA210 259 OC84 259; OX1873 209; OA201 509 OC16 509; OX1875 259; OX1873 209; OC26 409 OC171 309; DX1808 509; OC26 409 OC204 409; DX1808 509; OC26 409; OC204 409; DX1808 509; OC26 409; OC204 509; DX1818 276; OC28 609; OC204 759; DX1873 209; OC28 409; OC204 509; DX1873 209; OC38 459; IN43 109; SFR 5 329; OC48 129; IN670 79; SN128 879; OC46 129; IN670 71; SN128 879; OC46 129; IN670 71; SN128 879; OC46 129; IN670 71; SN128 879; OC470 129; IN700 71; SN128 879; OC48 129; IN670 71; SN128 879; OC49 129; IN670 71; SN128 879; OC40 129; IN670 71; SN148 979; ALL valves 10p GET115 45p 10p GET116 50p GEX66 £1-50 15p NKT222 20p 25p NKT304 50p 20p RA8310AF 913 84.25p t 19136 81.50p 8 19144 85.00p 81 19144 85.00p 80 12516GT 40p 906 30C15 75p 900 30C17 85p 9004 925 84p 9006 921 84p 9006 921 84p 9006 921 84p 9006 921 84p 9006 Individually packed VALVES quaranteed 6J5GT 6J6 6J7G £1 50p 20p 19G6 85n 19H4 B12H **\$1.75p** ECH84 **45p** N78 CY31 **35p** ECH200 **62p** GA2 DAF96 **38p** ECL80 **45p** OB2 DF96 **42p** ECL82 **35p** PAB £1.75p \$1.75p \$2.25p 20p 25p 50p 15p 15p 8J7M 5K6GT 6K7 6K7G 6K8GT 6K25 68A7 68A7GT 68C7GT 68G7 68J7 68J7 20p 20p 20p 45p 47p 83 p 26 p 31 p RITM CY31 DAF96 DF96 DK96 8D918 8D928 8D938 8D94 ECL80 ECL82 PARCSO 87 85p PABC80 65p PC97 42p PC900 45p PCC84 25p PCC89 PCC189 65p PCE800 25p PCF80 56p PCF82 B8 4/p 88729 25p 8D94 8B929 25p 8D94 8B 8D100 31-80p 8D98 40p 8D913 25p V405A 40p 87213 25p 2R21 33p 2R21 45p 7R22 42p 20p ECL86 42p ECL86 45p EF36 45p EF37A\$1.25p EF40 50p DI.09 25p AC127 25p AC128 64p AC176 1.50 AC17 25p 20p 20p DIOA DIOS 50p 55p 75p 80p 83p 57p 77p 48p 72p 65p | BYZ13 | 255 | V406A | 40p | 25p | BYZ16 | 63p | Z2A510(P 78p | 25p | BYZ16 | 63p | Z2A510(P 78p | 25p | CR81/10 | 25p | ZR11 | 83p | 25p | CR81/20 | 84p | ZR21 | 46p | 25p | CR81/30 | 40p | ZR22 | 42p | 25p | CR81/30 | 43p | 25p | CR83/30 | 30p | 25p | CR83/30 | 30p | 25p | CR83/40 | 25p | DIODES | 25p | CR83/40 | 50p | 42p | 25p | CR83/40 | 50p | 42p | 25p | CR83/40 | 50p | 10p | 25p | CR83/40 | 50p | 10p | 25p | CR83/40 | 25p | 10p | 25p | CR83/40 | 25p | CR83/40 | 25p | 25p | CR83/40 | 25p | CR83 PCC189 PCE800 PCF80 PCF82 PCF84 PCF86 PCF200 PCF201 C.R. 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6BC8			.48		24		- 55		. 20	EF92	
6BE6	- 80	6N7GT	.40		95	50L6GT	. 45			EF98	
6BH6	.48	6Q7GT	.48	128G7	. 28		.48			EF183	
6BJ6		6Q7M		128H7	.15		40	EBC81	99	EF184	
6BQ7A				128J7		90C1				EH90	
9 DQ / A	. 66	191610	. 60	1 1 400 1	. 20	- 20CI	.50	PDIO	. 00	DIESO	_
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	EL32 ·18	KTW62 .68	PL505 1.30	UL84 .2	28 l	AC156	.20		.20	OA81	.09
	EL34 .44	KTW63 .50	PL508 .90	UM80 .8		AC157	. 25	BCY34	.23	OA85	.08
	EL35 1.00	MHLD6 .75	PL509 1.80	UY41 .8			.25	BCY38	.28	OA90	.18
	EL37 .74	P61 .44	PL802 .75	UY85 .8			.25		. 25	OA91	.09
	EL41 .58	PABC80 .32	PM84 .81	U10 .4			.38	BCZ11	. 38	OA95	.09
		PC86 .44	PX4 1.16	U12/14 .8			. 55	BF158	29	OA200	.09
7.1	EL42 .58	PC88 .44	PY33/2 .50	U19 1.7			.28	BF159	.25	OA202	.10
	EL81 .50		PY80 .38	U22 .8			. 25	BF163	.20	OC22	.88
90	EL83 .88	PC95 .58					.20	BF173	.38	OC23	.38
	EL84 .21	PC97 .36	PY81 .24				.19	BF180	.80	OC24	.38
. 88	EL85 .40	PC900 .29	PY82 .23				.18	BF181	.40	OC25	.88
.26	EL86 .38	PCC84 .27	PY83 .26	U45 .7			.19	BF185	.40	OC28	.60
. 60	EL91 .23	PCC85 .24	PY88 .81	U52 8			.15	BF194	. 15	OC29	.63
.59	EL95 .32	PCC88 .89	PY500 .95	U191 .		ACY28	.18	BFY50	.23	OC35	.82
.59	ELL80 .75	PCC89 .42	PY800 .81	U251 .6		AC 1 20	.36	BFY51	.19	OC36	.48
. 34	EM80 .87	PCC189 .46	PY801 .81		10			BFY52	.20	OC42	.68
1.50	EM81 .37	PCF80 .26	PZ30 .48	U403 .8	33	AD149	.50	BY100		OC44	10
1.50	EM83 .75	PCF82 .80	QQV03/10	U404 .8	58		. 45	BY105	.18	OC45	:11
.60	EM84 .81	PCF84 .40	1.20	U801 .7			.45	BY114	.18	OC46	.15
.16	EM85 1.00	PCF86 .44	Q895/10 .49	U4020 .8		ADT140	.03	BY126	.15	OC70	.18
.19	EM87 .84	PCF200 .67	QV04/7 .63	VP23 .4			. 50	BY 127	.18	OC71	.11
. 21	EY51 .29	PCF801 .28	R10 .75	VP41 .2		AF114	. 25 . 15	BYZ10	. 25	OC72	:iil
- 28	EY81 .35	PCF802 .87	R11 .98	VT61A .				BYZ11	. 25	0C74	.28
. 88	EY83 .54	PCF806 .55	R16 1.75	VT501 ·1		AF117	.19	BYZ12	. 25	OC75	.11
. 35	EY84 .50	PCH200 .68	TPT1 - 000	VU111 .4		AF121	.25	BYZ13	. 25	OC76	.15
. 48	EY87/6 .27	PCL82 .29	R19 .28		30	AF124 AF125	.17	CG12E	.20	0C77	.27
	EY88 .40	PCL83 .54	8P61 .33	VU120A .6			.18	F8Y11A		OC78	.15
1.70	EZ40 .40	PCL84 .82	TH233 .98	VU133 .8			. 65	FSY41A		OC78D	.15
.27	EZ41 .42	PCL805/85	TP2620 .98	W107 .8 W729 .6			. 68	GD9	. 20	OC81	.11
.25	EZ80 .19	.87	UABC80.80 UAF42.49		50	AF180	. 48	GET113		OC81 D	iii
.64	EZ81 .20	PCL86 .86	UAF42 .49 UBC41 .45	Transistor		AF186	. 55	GET118		OC82	iii
	FW4/500.75	PD500 1.44	UBC81 .40	& Diodes	•	AF239	. 38	GET119		OC82D	iii
2.10	FW4/800.75	PEN4DD	UBF80 .28		18	BA102	. 45	GET573		OC83	.20
. 63	GY501 -75	1.38	UBF89 .28	2N2297		BA115	.14	GET587		OC84	.24
. 57	GZ30 .38	PEN45DD	UBL21 .55	2N2369 .		BA116	.25	GET873		OC123	.23
. 25	GZ32 .89	PEN453DD	UC92 .85	2N 3053 .8		BA129	.13	GET887		OC139	.23
.88	GZ33 .70	. 98	UCC84 .88	2N3121 2-5		BA130	.10	GET897		OC140	.95
.84	GZ34 .47 GZ37 .67	PENDD-	UCC85 .38	2N3709 .		BA153	.15	GET898		OC169	.28
. 28	GZ37 .67	4020 .88	UCF80 .31	2N3988 .		BC107	.18	MI	.15	OC172	.35
.28	HABC80.44 HL23DD.40	PEN45 .40	UCH21 .60	28323 .		BC108	.13	OA5	. 28	OC200	.22
.54	HL41DD.98	PEN46 .20	UCH42 .57		15	BC113	.25	OA9	.18	OC201	.38
. 54	HL42DD.50	PFL200 .50	UCH81 .29		15	BC115	.15	OA10	. 43	OC202	.48
. 33	HN309 1-40	PL33 .38	UCL82 .30		15	BC116	. 25	OA47	.10	OC203	.80
. 68	HVR2 .58	PL36 .46	UCL83 .48		18	BC118	. 23	OA70	.15	OC204	.80
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.58	KT2 .25	PL81A .48	UF42 .60		25	BCY12	.50	OA79	. 09	ORP12	. 58
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5mA £2-80	20 amp. A.C. 22.80
10mA £2:80	30 amp. A.C. 22.80
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50-0-50µA £2-60	20 V. D.C £2-00
100µA £2:60	50V. D.C #2-00
100-0-100µA £2-50	300V, D.C. #2.00
500μA £2:30	15 V. A.C £2-10
1mA £2 00	300V. A.C 22-10
5mA £2.00	8 Meter 1m A #2-10
10mA £2:00	VU Meter £3.20
50mA #2:00	1 amp. A.C. \$2.00
100mA #2·00	5 amp. A.C. #2-00
500mA 42:00	10 amp. A.C. \$2.00
1 amp \$2.00	20 amp. A.C. * 22.00
5 amp £2.00	30 amp. A.C. \$2.00

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20V. D.C #2-20
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150V. D.C. 22-20
300V. D.C. #2-20
15V. A.C £2:30
50V. A.C #2:30
150V. A.C 22-30
300V. A.C £2.30
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VU Meter #3-37
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100mA A.C. #2:20
200mA A.C. #2.20
500mA A.C. #2-20
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100μA £1 90	15V. D.C \$1-60					
100-0-100µA \$1.75	20V. D.C £1-60					
200μA £1.75	100 V. D.C. 21-60					
500μA 21.65	150V. D.C. #1-60					
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1mA £1.60	500V. D.C. 21-60					
1-0-1mA #1-60	750V. D.C. 21-60					
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100mA £1.60	S Meter 1mA 21.70					
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			8# - PG114	15p, BFW90 22p, NKT219 30p	Integrated FJH111 70p 8N7437 64p	VALVES
Transistors 2G301 20p	2N3415 2N3416 2N3417	22p 2N5458 37p 2N5459 37p 28102	35p BC114 40p BC115 25p BC116	15p BFW91 20p NKT223 27p 15p BFX12 22p NKT224 22p	Circuits FJH121 25p 8N7440 20p FJH131 25p 8N7441AN	OA2 38p125Z4 30p1EM80 45p
2G302 20p 2G303 20p	2N3439	130p 28103 97p 28104	25p BC118 25p BC119	15p BFX13 22p NKT225 22p 30p BFX29 25p NKT229 30p	CA3000 180p FJH141 25p 75p CA3005 117p FJH151 25p SN7442 75p	OB2 45p 25Z5 42p EM81 60p OZ4 80p 25Z6 65p EM84 85p IL4 20p 30C15 80p EM85 \$1.00
2G306 42p 2G308 30p	2N3564 2N3565	17p 28301 15p 28302	50p BC121 50p BC122	20p BFX30 25p NKT237 35p 20p BFX37 30p NKT238 25p	CA3007 262p FJH161 70p 8N7446 190p CA3011 75p FJH171 25p 8N7447 135p CA3012 88p FJH181 25p 8N7448 125p	IL4 20p 30C15 80p EM85 \$1.00 IR5 40p 30C17 90p EM87 70p IS5 30p 30C18 80p EY51 40p
2G309 30p 2G371 15p	2N3566 2N3568	22p 28303 25p 28304	60p BC125 75p BC126	15p BFX44 87p NKT240 27p 20p BFX68 67p NKT241 27p	CA3013 106p FJH221 25p SN7450 20p	1T4 25p 30F5 85p EY86 40p 1U4 30p 30FL1 75p EY87 42p
2G374 20p 2G381 22p	2N3569 2N3570	25p 28501 125p 28502	32p BC134 35p BC135	12p BFX84 25p NKT242 20p 12p BFX85 30p NKT243 52p	CA3018 84p FJH241 25p SN7453 20p	1U5 60p 30FL12 120p EZ40 55p 2D21 85p 30FL14 95p EZ41 50p
2N388A 49p 2N404 20p	2N3572 2N3605	97p 28503 27p 3N83	27p BC136 40p BC137	15p BFX86 25p NKT244 17p 15p BFX87 25p NKT245 20p	110p FJJ101 50p SN7460 20p	3Q4 50p 30L15 85p EZ80 27p 384 35p 30L17 80p EZ81 29p
2N696 15p 2N697 15p	2N3607	27p 3N128 22p 3N140	70p BC138 77p BC140	20p BFX88 20p NKT261 20p 35p BFX89 62p NKT262 30p	CA3020 126p FJJ121 60p SN7473 40p	3V4 48p 30P12 80p GZ32 48p 5R4 75p 30P19 85p GZ34 60p
2N698 25p 2N699 30p	2N3638A		72p BC141 55p BC147	85p BFX93A 70p NKT264 20p 10p BFY11 42p NKT271 20p 10p BFY18 25p NKT262 20p	CA3020A FJJ131 60p 8N7474 40p 160p FJJ141 125p 8N7475 45p CA3021 156p FJJ181 75p 8N7476 45p	5U4 85p 30PL1 75p KT66 22.05 5V4 45p 30PL13 93p KT88 22.00
2N706 10p 2N706A 12p	2N 3642	18p 3N143 18p 3N152	67p BC148 87p BC149	12p BFY19 25p NKT274 20p	CA3022 130p FJJ191 65p SN7483 87p CA3023 128p FJJ211 125p SN7486 33p	5Y3 40p 30PL14 90p MU14 75p 5Z4G 40p 35L6 50p PABC80 40p
2N708 15p 2N709 62p	2N3644	20p 40050 25p 40250	55p BC152 50p BC153	20p BFY24 45p NKT278 25p	CA3026 100p FJJ251 125p SN7490 87p CA3028A 74p FJL101 125p SN7492 87p	6/30L2 80p 35W4 85p PC86 60p 6AC7 40p 35Z4 85p PC88 60p
2N718 25p 2N718A 30p	2N3691	25p 40251 15p 40309	82p BC154 82p BC157	15p BFY30 40p NKT401 87p	CA3028B FJY101 25p 8N7493 87p 105p IC12 \$1.80 8N7495 87p	6AG7 40p 35Z5 50p PC97 45p 6AK5 85p 50B5 50p PC900 48p
2N726 80p 2N727 80p	2N 3693	18p 40310 15p 40311 18p 40312	45p BC158 35p BC159 47p BC160	11p BFY41 50p NKT402 90p 12p BFY43 62p MKT403 75p 85p BFY50 20p NKT404 35p	CA3029 87p L900 40p SN7496 87p CA3029A L914 40p SN74107 52p	6AK6 60p 50C5 50p PCC84 40p 6AL5 20p 80 55p PCC85 40p
2N914 17p 2N916 17p	2N3702	10p 40314	87p BC167	11p BFY51 20p NKT405 75p 10p BFY52 20p NKT406 82p	165p L923 40p 8N74153 CA3030 187p LM380 128p 185p	6AM6 30p 85A2 50p PCC88 55p 6AQ5 38p 807 50p PCC89 50p
2N918 30; 2N929 22; 2N930 20;	2N3704	11p 40316	47p BC168C 87p BC169B	11p BFY53 15p NKT451 62p 11p BFY56A 57p NKT452 62p	CA3035 122p MC724P 60p 8N74154 CA3036 72p MC780P 247p 200p	6A86 40p 1625 50p PCC189 55p 6AT6 35p 5763 70p PCF80 30p
2N930 20p 2N987 40p 2N1090 22p	2N3706	10p 40317 9p 40319 11p 40320	55p BC169C 47p BC170	12p BFY76 42p NKT453 47p 12p BFY77 57p NKT713 20p	CA3039 82p MC788P 146p SN74160 CA3041 109p MC790P 124p 180p	6AU6 25p 6146 160 PCF82 34p 6AV6 30p AZ31 55p PCF84 60p
2N1090 221 2N1091 221 2N1131 251	2N3708	7p 40323 9p 40324	82p BC171 47p BC172	15p BFY90 65p NKT717 42p 15p B8X19 17p NKT734 27p	CA3042 109p MC792P 66p SN74161 CA3043 187p MC799P 66p 260p	6BA6 25p CY31 35p PCF86 60p 6BE6 30p DAF91 30p PCF800 80p
2N1132 251 2N1302 171	2N 3710	9p 40326 12p 40329	37p BC175 30p BC177	22p B8X20 15p NKT736 35p 20p B8X21 20p NKT733 25p	CA3044 120p MC1303L SN74164 CA3045 122p 200p 220p	6BH6 76p DAF96 45p PCF801 50p 6BJ6 50p DF91 32p PCF802 50p
2N1303 17; 2N1304 22;	2N3713 2N3714	187p 40344 200p 40347	27p BC178 57p BC179	20p B8X26 45p NKT781 30p 20p B8X27 47p OC16 50p	CA3046 81p MC1304P 8N74165 CA3047 137p 225p 225p CA3048 204p MC1305P 8N74192	6BQ7A 40p DF96 45p PCF805 80p 6BR7 90p DK91 40p PCF806 70p 6BR8 70p DK92 55p PCF808 75p
2N1305 221 2N1306 251	2N3715 2N3716	123p 40348 130p 40360	52p BC182 40p BC182L	10p B8X28 82p OC19 87p 10p B8X60 82p OC20 75p	CA3049 160p 886p 175p	6BR8 70p DK92 55p PCF808 75p 6BW6 85p DK96 50p PCL82 35p 6BW7 80p DL92 35p PCL83 65p
2N1307 251 2N1308 251	2N3773 2N3791	240p 40361 206p 40362	40p BC183 50p BC183L	9p BSX61 62p OC22 50p 9p BSX76 15p OC23 60p	CA3050 185p MC838P 8N74193 CA3051 184p 549p 175p CA3052 165p MC1435P TAA241	6BZ6 40p DL94 48p PCL84 45p 6C4 33p DL96 45p PCL85 40p
2N1309 251 2N1507 171	2N 3820	84p 40370 55p 40406 50p 40407	32p BC184 57p BC184L 40p BC186	11p B8X77 20p OC24 60p 11p B8X78 25p OC25 40p 25p B8Y24 15p OC26 25p	CA3053 46p 845p 162p CA3054 109p MC1552G TAA242	6CD6 125p DM70 40p PCL86 45p 6CL8 50p DY86 32p PFL200 65p
2N1613 201 2N1631 851	2N3854	27p 40408	40p BC186 52p BC187 55p BC212L	27p B8Y25 15p OC28 60p 12p B8Y26 17p OC29 60p	CA3055 240p 461p 425p CA3059 165p MC1709CG TAA243 150p	6CW4 65p DY87 33p PL36 55p 6F1 62p E88CC 100p PL81 50p
2N1632 801 2N1637 801	2N3855	27p 40410	62p BC213L 50p BC214L	12p BSY27 15p OC35 50p 15p BSY28 17p OC36 60p	CA3064 120p 94p TAA263 75p FCH101 85p MFC4000P TAA293 97p	6F6G 35p E180F 100p PL82 45p 6F13 45p EABC80 35p PL63 45p
2N1638 27; 2N1639 27; 2N1701 169;	2N 3856	80p 40467A	57p BCY10 85p BCY30	27p BSY29 17p OC41 22p 27p BSY32 25p OC42 25p	FCH111 105p 55p TAA300 175p FCH121 105p 8N7400 20p TAA310 125p	6F14 70p EAF42 85p PL84 40p 6F15 65p EB91 20p PL500 75p
2N1711 241 2N1889 321	2N3858	25p 40528	72p BCY31 57p BCY32	30p BSY36 25p OC44 15p 50p BSY37 25p OC45 12p	FCH131 50p SN7401 20p TAA320 72p FCH141 105p SN7402 20p TAA350 175p	6F18 50p EBC41 55p PL504 80p 6F23 65p EBC81 80p PY32 55p
2N1893 37 2N2147 72	2N3859	27p 40603	50p BCY33 30p BCY34	25p BSY38 20p OC46 15p 30p BSY39 22p OC70 15p	FCH151 105p SN7403 20p TAA435 147p FCH171 105p SN7404 20p TAA521 132p	6H6 17p EBF80 40p PY33 68p 6J4 50p EBF83 40p PY80 40p
2N2160 57 2N2193 40	2N3860	80p AC126	20p BCY38 24p BCY39	40p B8Y43 50p OC71 12p 60p B8Y51 32p OC72 12p	FCH181 105p 8N7400 20p TAA522 300p FCH191 105p 8N7406 80p TAA530 495p	6J5 25p EBF89 32p PY81 30p 6J5GT 30p EBL21 60p PY82 35p
2N2193A 48 2N2194 27	2N3877	40p AC128	20p BCY40 18p BCY41	50p BSY52 32p OC73 30p 15p BSY53 37p OC74 30p	FCH201 180p 8N7408 20p TAA811 445p FCH211 180p 8N7409 20p TAB101 97p FCH221 180p 8N7410 20p TAD100 150p	6J6 20p EC86 60p PY83 88p 6J7 45p EC88 60p PY88 40p 6KRG 40p ECC40 65p PY800 40p
2N2194A 80 2N2217 85	2N3900 2N3900		22p BCY42 22p BCY43	15p BSY56 40p OC75 22p 15p BSY56 90p OC76 22p	FCH231 150p SN7411 23p TAD110 150p	6L6GT 45p ECC84 80p PY801 50p
2N2218 20 2N2219 20	2N3903	97p AC176 20p AC187	20p BCY54 25p BCY58	32p B8Y79 45p OC77 30p 22p B8Y90 57p OC78 20p	FCJ101 160p 8N7412 48p 8L403D 16 p FCJ111 150p 8N7413 30p 8L702C 147p FCJ121 275p TE7416 84p UA702A 280p	6LD20 50p ECC85 40p U25 80p 6Q7 40p ECC88 40p U26 80p 68A7 40p ECF80 35p U50 40p
2N2220 85 2N2221 25	2N3905	25p AC188 80p ACY17	25p BCY59 27p BCY60	22p BSY95A 12p OC81 20p 97p C424 15p OC81D 20p 15p C450 15p OC82 25p	FCJ131 275p SN7417 84p UA702C 77p FCJ141 895p SN7420 20p UA703C 187p	68G7 40p ECF82 85p U52 85p 68J7 40p ECF86 65p U191 75p
2N2222 20 2N2222 A 25 2N2297 30	2N4058	25p ACY18 12p ACY19 10p ACY20	24p BCY70 24p BCY71 20p BCY72	15p C450 15p OC82 25p 20p GET102 30p OC82D 15p 15p GET113 20p OC83 25p	FCJ201 100p 8N7423 51p UA709C 45p FCJ211 275p 8N7427 48p UA710C 125p	68K7 40p ECH21 57p U281 40p 68L7 25p ECH35 100p U282 40p
2N2297 80 2N2368 15 2N2369 15	2N4060	12p ACY21 12p ACY22	20p BCY78 10p BCY79	30p GET114 20p OC84 25p 30p GET118 20p OC139 25p	FCK101 430p SN7428 80p UA716 187p FCL101 230p SN7430 20p UA723C 100p	68N7 35p ECH42 75p U301 40p 68O7 40p ECH81 30p U801 \$1.80
2N2369A 15 2N2410 42	2N4062	12p ACY28 47p ACY39	17p BCZ10 47p BCZ11	27p GET120 25p OC140 32p 40p GET873 12p OC170 25p	FCY101 102p SN7432 48p UA730C 160p FJH101 25p SN7433 80p UA741C 80p	6U4 6bp ECH83 45p UABC80 40p 6V6G 25p ECL80 45p UAF42 55p
2N2483 27 2N2484 32	2N 4248	15p ACY40 15p ACY41	20p BD112	50p GET880 80p OC171 80p 112p GET887 20p OC200 40p	BRIDGE 50 PIV 4A 40p RECTIFIERS 100 PIV 4A 50p	6V6GT 32p ECL82 35p UBC41 50p 6X4 35p ECL83 70p UBC81 40p
2N2539 22 2N2540 22	2N4250	18p ACY44 42p AD140	25p BD121 47p BD123	65p GET889 22p OC201 60p 80p GET890 22p OC202 75p	PLASTIC 200 PIV 4A 55p	6X5G 80p ECL86 40p UBF80 40p 6X5GT 40p EF37A 120p UBF89 85p 10C2 50p EF39 50p UCC84 49p
2N2613 85 2N2614 80	2M4284	17p AD149	47p BD124 62p BD131	60p GET896 22p OC203 40p 75p GET897 22p OC204 40p	ENCAPSULATED 600 PIV 4A 70p 600 PIV 1A 50p 50 PIV 6A 45p	10F1 75p EF40 50p UCC85 40p
2N2646 40 2N2711 25	n 2N4286	17p AD161 17p AD162		80p GET898 22p OC205 75p 125p MAT100 25p OC206 95p 105p MAT101 25p OC207 75p	50 PIV 2A 45p 100 PIV 6A 55p 100 PIV 2A 50p 200 PIV 6A 65p	10P13 60p EF41 65p UCF80 55p 10P14 \$1.10 EF42 70p UCH21 60p 12AT6 80p EF80 25p UCH42 70p
2N2712 25 2N2713 27 2N2714 30	p 2N4288	17p AF109 15p AF114 17p AF115	45p BDY20 25p BDY61 25p BDY62	125p MAT120 25p OCP71 42p	200 PIV 2A 55p 400 PIV 6A 75p 400 PIV 2A 60p 600 PIV 6A 85p	12AT7 30p EF85 35p UCH81 40p 12AU7 30p EF86 30p UCL82 35p
2N2714 80 2N2904 20 2N2904A 25	2N4290	12p AF116 15p AF117	25p BF115 20p BF117	85p MJ400 107p ORP60 40p 47p MJ420 80p ORP61 42p	SILICON RECTIFIERS MINIATURE WIRE ENDED PLASTIC	12AX7 80p EF89 28p UCL83 60p 12AV6 40p EF91 80p UF41 60p
2N2905 25	p 2N4292 p 2N4294	15p AF118 17p AF121	60p BF152 80n BF154	88p MJ421 80p P346A 22p 90p MJ430 102p ST140 15p	SERIES IN PL CL 1 AMP 1.5 AMP 8 AMP	12BA6 40p EF92 35p UF80 35p 12BE6 40p EF183 35p UF85 40p
2N2906 20 2N2906 A 25	p 2N4303 p 2N4964	47p AF124 15p AF125	22p BF158 19p BF159	15p MJ440 95p ST141 20p 85p MJ480 97p TIS34 62p	4001 50PIV 7p 8p 19p 4002 100PIV 7p 9p 20p	12BH7 45p EF184 85p UF89 40p 19AQ5 85p EH90 40p UL41 65p
2N2907 28 2N2923 15	p 2N4965 p 2N5027	18p AF126 52p AF127	19p BF163 16p BF167	85p MJ481 185p TIS43 40p 18p MJ490 100p TIS44 10p	4003 200PIV 8p 10p 22p 4004 400PIV 8p 10p 25p	20D1 50p EL34 50p UL84 40p 20F2 65p EL41 60p UY41 48p 20L1 41:10 EL42 65p UY85 40p
2N2924 15 2N2925 15	p 2N5028 p 2N5029	57p AF139 47p AF178	28p BF170 42p BF173	88p MJ491 187p TIS45 27p 18p MJE340 50p TIS46 11p	4005 600 PIV 10p 12p 26p 4006 800 PIV 12p 15p 27p 4007 1000 PIV 15p 16p 30p	20P1 50p EL81 55p VR105/30 38p
2N2926G 10 2N2926O 10	p 2N5030 p 2N5172	42p AF179 12p AF180	45p BF177 50p BF178	80p MJE370 80p TI847 11p 85p MJE371 80p TI848 12p 80p MJE520 60p TI849 12p	50 + less 15 % 100 + less 20 %	20P3 60p EL84 25p VR150/30 80p 20P4 \$1.10 EL85 43p 20P5 \$1.20 EL91 35p Add 12p in \$
2N2926Y 10 2N3011 20	p 2N5175	52p AF181 52p AF186 45p AF239	40p BF179 39p BF180 30p BF181	80p MJE520 60p TI849 12p 85p MJE521 70p TI850 12p 82p MPF102 42p TI851 10p	SILICON RECTIFIERS STUD MOUNTING	25L6 50p EL96 35p for postage
2N3014 88 2N3053 18 2N3054 46	p 2N5232	A 80p AF279	47p BF182 47p BF184	30p MPF103 35p TIS52 11p 20p MPF104 37p TIS53 22p	6A 10A 17.5A 35A 100PIV — 45p 50p \$1.22	DIODES & RECTIFIERS 1N34A 10p BA154 12p GJ7M 87p 1N914 7p BAX13 12p OA5 17p
2N3055 60 2N3133 80	p 2N5246	42p AFZ11	32p BF185 25p BF194	20p MPF105 87p XB112 12p 15p MPS3638 32p XC141 35p	200 PIV 25p 50p 55p \$1.42 400 PIV 30p 55p 62p \$1.77 600 PIV 32p 60p 72p \$2.12	1N916 10p BAX16 7p OA6 18p
2N3134 18 2N3135 28	p 2N5265 p 2N5305	325p ASY27 37p ASY28	30p BF195 24p BF196	15p NKT124 42p ZTX107 15p 15p NKT125 27p ZTX108 12p	600PIV 82p 60p 72p £2:12 800PIV 85p 75p 87p £2:47 1000PIV 40p 85p £1:05 £2:77	AA 100 10 BA 728 150 CA9 100
2N3136 20 2N3390 20	p 2N5306 p 2N5307	40p A8Y29 87p A8Y50	27p BF197 25p BF198	15p NKT126 27p ZTX109 15p 15p NKT128 27p ZTX300 12p	50 + less 15% 100 + less 20%	AAZ15 10p BY103 22p OA70 7p BA100 16p BY122 87p OA73 10p
2N3391 20 2N3391A 30	2N5308 p 2N5309	87p ASY51 62p ASY54	32p BF200 25p BF224	35p NKT135 27p ZTX301 15p	ZENER DIODES	BA102 80p BY124 15p OA79 7p BA110 25p BY126 12p OA81 38p
2N3392 17 2N3393 1	p 2N5310 p 2N5354	42p ASY67 27p ASY86	45p BF225 32p BF237	19p NKT210 80p ZTX303 20p 23p NKT211 80p ZTX304 25p	3·3·33 V 2·4—100 3·9—100V 10p each 25p each 40p each	■ D. 130 #6- D37144 K9s O.490 78
2N3394 13 2N3402 23	p 2N5355 p 2N5356	82p AUY10		22p NKT212 80p ZTX500 15p 83p NKT213 80p ZTX501 15p 47p NKT214 20p ZTX502 20p	25 + less 15% 100 + less 20% TRANSISTOR DISCOUNTS:- 12 + 10%;	BA115 7p BY210 85p OA91 7p BA141 82p BYZ11 30p OA95 7p
2N3404 8	2N5365 2N5366	32p BC108	10p BFW61 10p BFW87 10p BFW88	25p NKT215 22p ZTX503 17p	25 + 15%; 100 + 20% any one type. Postage on all Semi Conductors 7p extra.	BA142 82p BYZ12 80p OA200 7p BA144 12p BYZ13 25p OA202 10p
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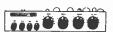
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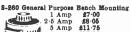
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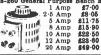
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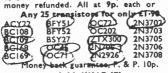
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184 0.8		6F25 1.00	10P13 0.60	FIEC	TRON	IC VA	LVES	EL36	0.50	PC86 0.60	PL508 0.90	
185 0.8				ELEU	INON	10 17	_ ,	EL37	1.60	PC88 0.60	PL509 1.10	UBC41 0.50
1T4 0.8		6F26 0.35		0.55441 0.80	50C5 0.50	DK96 0.50	ECC83 0.30	EL41	0.60	PC92 0.05	PL801 0.80	UBC81 0.40
1U4 0.8		6F28 0.65	12AB5 0.70	25Z4G 0.30	50CD6G1.20	DK92 0.55	ECC84 0.30	EL81	0.55	PC97 0.50	PL802 0.95	UBF80 0.40
1U5 0.6	D 6BF6 0.55	6GK6 0.60	12AC6 0.50	25Z6GT 0.65		DK96 0.60	ECC85 0.40	EL83	0.42	PC900 0.48	PM84 0.60	UBF89 0.35
1V2 0.5	6BH6 0.75	6J4 0.50	12AD6 0.55	30 A 5 0.50	50EH5 0.60		ECC88 0.40	EL84	0.25	PABC800-40	PY31 0.30	UBL1 0.60
1X2B 0.5	6BJ6 0.50	6J5GT 0.80	12AE6 0.55	30 A E 3 0.40	50L6GT 0.55	DL96 0.45		EL85	0.48	PCC84 0.40	PY33 0.68	UBL21 0.65
2A3 0.5		6J6 0.20	12AL5 0.50	30C1 0.80	85 A 2 0.50	DM70 0.55	ECC89 0.50			PCC85 0.40	PY80 0.40	UC92 0.40
2CW4 0.7		6J7 0.45	12AQ5 0.50	30C15 0.80	90 AG 2.40	DM160 0.65	ECC91 0.20	EL86	0.40		PY81 0.30	UCC85 0.40
		6K6GT 0.60	12AT6 0.80	30C17 0.90	90 A V 2.50	DY51 0.50	ECC189 0.65	EL90	0.88	PCC88 0.55	PY82 0.35	UCF80 0.55
		6K7 0.35	12AT7 0.85	30C18 0.80	90C1 0.65	DY86 0.82	ECC807 1.25	EL95	0.35	PCC89 0.55		UCH21 0.60
3A4 0.4		6K8G 0.40	12AU6 0.35	30F5 0.85	90CV 2.40	DY87 0.83	ECF80 0.35	EL821	0.60	PCC189 0.55	PY83 0.38	UCH42 0.70
3A5 0.7			12AU7 0.80	30FL1 0.75	807 0.50	DY802 0.85	ECF82 0.35	EL822	1.80	PCC805 0.85	PY88 0.40	
3BP1 8.0				30FL12 1.20	813 8.75	E88CC 0.65	ECF86 0.65	ELL80	0.75	PCC806 0.80	PY500 1.00	UCH81 0.40
384 0.8		6L6GT 0.45			866A 0.75	E180F 1.00	ECF8041.65	EM34	1.00	PCF80 0.80	PY800 0.40	UCL81 0.60
5R4GY 0.7	5 6BW7 0.80	6L7 0.40	12AV7 0.65	30FL14 0.85	5642 0.70	EABC800.35	ECH42 0.75	EM71	0.80	PCF82 0.35	PY801 0.50	UCL82 0.35
5U4G 0.8	5 6BX6 0.25	6L18 0.45	12AX7 0.80	30L1 0.40		EAF42 0.55	ECH81 0.30	EM80	0.45	PCF84 0.60	PZ30 0.35	UCL83 0.60
5V4G 0.4	5 6BZ6 0.40	6LD20 0.50	12AY7 0.75	30L15 0.85	5670 0.50		ECH83 0.45	EM81	0.60	PCF86 0.60	QQVQ2-6	UF9 0-60
5Y3GT 0.4	0 6C4 0.83	6N7GT 0.45	12B4A 0.60	30L17 0.80	6080 1.60	EAF8010.50	ECH84 0.45	EM83	0.50	PCF87 0.90	2.25	UF11 0.50
5Z3 0.6		6P1 0.60	12BA6 0.40	30P12 0.80	6146 1.60	EBC33 0.50	ECL80 0.45	EM84	0.85	PCF801 0.50	QQVO3-10	UF41 0.60
5Z4G 0.4		6P28 0.65	12BA7 0.40	30P19 0.85	6146B 2.50	EBC41 0.55		EM85	1.00	PCF802 0.50	1.25	UF42 0.60
6/30L2 0.8		6Q7 0.40	12BE6 0.40	30PL1 0.75	6360 1.25	EBC81 0.80	ECL81 0.50		0.70	PCF805 0.80	QQV03-20A	UF43 0.60
6AB4 0.3		68A7 0.40	12BH7 0.45	30PL13 0.98	6939 2.25	EBF80 0.40	ECL82 0.85	EM87		PCF806 0.70	5.25	UF80 0.35
6AF4A 0.5		68G7 0.40	12BY7 0.60	30PL14 0.90	7199 0.80	EBF83 0.40	ECL83 0.70	EN91	0.88	PCF808 0.70	QQV06-40A	UF85 0.40
		68K7 0.40	12E1 8.00	35 A3 0.55	7360 2.00	EBF89 0.80	ECL84 0.55	EY51	0.40		5.50	UF89 0.40
6AG5 0.2	6CW4 0.65	68L7GT0.35	12K5 0.70	35A5 0.75	7586 1.25	EBL31 1.50	ECL85 0.55	EY80	0.55	PCH2000.70		UL84 0.40
6AG7 0.4		68N7GT0.85	12K7GT0.40	35B5 0.65	7895 1.25	EC53 0.50	ECL86 0.40		0.40	PCL81 0.50		UM84 0.25
6AH6 0.5	6CY5 0.45		1207GT0.40	35C5 0.50	9002 0.40	EC86 0.60	ECLL800	EY83	0.55	PCL82 0.85	U18/20 0.75	
		68Q7 0.40		35D5 0.75	A2293 2.00	EC88 0.60	2.00	EY86	0.40	PCL83 0.65	U25 0.80	UYIN 0.50
6AJ8 0.8		68R7 0.40	198R7 0.40		AZ1 0.55	EC90 0.88	EF40 0.50		0.48	PCL84 0.45	U26 0.80	UY11 1.00
6AK5 0.8		6T8 0.85	20D1 0.50	35L6GT0.50	AZ31 0.55	EC92 0.35	EF41 0.65		0.48	PCL85 0.40		UY41 0.48
6AK6 0.6	6DK6 0.50	6U4GT 0.65	20L1 T.10	35W4 0.85		EC93 0.55	EF42 0.70		0.50	PCL86 0.45	U37 2.10	UY82 0.50
VAR-001	6DQ6B 0.70	6U8A 0.40	20P1 0.50	35Z3 0.70	CBL1 0.90		EF80 0.25		0.50	PCL88 0.90	U52 0.35	
	01004 1-20	6V6GT 0.40	20P4 1.10	35Z4G 0.85	CBL31 1.00		EF83 0.55		0.27	PCL800 0.98		
6AL5 0.9	6EA8 0.60	6X4 0.85	20P5 1.20	35Z5GT 0.60	CY31 0.85		EF85 0.35		0.29	PCL801 0.75		Z759 8-00
6AM5 0.8		6X5GT 0.40	25C5 0,50	50A5 0.75	DAF96 0.45	ECC81 0.35			8.00	PD500 1.30		
6AM6 0.3		6X8 0.60	25L6GT 0.50	50 B5 0.50	DF96 0.45	ECC82 0.80	EF86 0.30	, GIIC	0.00	1 1 1000 1.00		
OAMO U.C	0.10.											

TRANSISTORS 0.20 0.15 0.35 0.20 BD123 BD131 BD132 BF115 AC127 AC128 AC132 AC153

2N 705	0.70	ACIOS	0.20	201110	0.20
2N706	0.10	AC154	0.15	BF167	0.18
2N708	0.16	AC157	0.20	BF173	0.20
2N753	0.25	AC169	0.10	BF179	0.30
2N929	0.22	AC176	0.25	BF180	0.85
2N930	0.25	AC187	0.80	BF181	0.25
2N987	0.80	AC188	0.80	BF184	0.25
2N1131	0.25	ACY17	0.275	BF185	0.20
2N1132	0.25	ACY18	0.20	BF194	0.15
2N1184	1.25	ACY19	0.25	BF195	0.15
2N1301	0.40	ACY20	0.20	BF196	0.20
2N1302	0.75	ACY21	0.20	BF197	0.20
2N1304	0.25	ACY22	0.15	BF200	0.35
2N1305	0.25	AD140	0.50	BFW87	0.25
2N1306	0.25	AD149	0.50	BFW88	0.23
2N1307	0.80	AD161	0.35	BFW89	0.20
2N1308	0.25	AD162	0.35	BFW91	0.20
2N1309	0.80	ADZ11	1.25	BFX 88	0.25
2N1613	0.22	ADZ12	1.25	BFY17	0.40
2N1711	0.25	AF114	0.20	BFY19	0.60
2N1756	0.75	AF115	0.20	BFY50	0.25
2N2147	0.75	AF116	0.20	BFY51	0.20
2N2160	0.65	AF117	0.20	BFY52	0.25
2N2217	0.80	AF118	0.45	BSY26	0.20
2N2218	0.80	AF125	0.25	B8Y27	0.20
2N2219	0.85	AF127	0.20	B8Y28	0.20
2N2369A	0.20	AF180	0.85	BSY65	0.20
2N2477	0.65	AF181	0.85	BSY95A	0.15
2N2646	0.60	AF186	0.40	OC16	0.50
2N2905	0.85	AF239	0.40	OC22	0.50
2N2923	0.15	AFZ11	0.45	OC23	0.60
2N2924	0.15	ASY26	0.25	OC24	0.60
2N2926	0.125	A8 Y27	0.80	OC25	0.85
2N3053	0.25	A8Y28	0.25	OC26	0.25
2N3054	0.60	A8Y29	0.80	OC28	0.60
2N3055	0.75	ASY54	0.25	OC29	0.60
2N3133	0.30	ASZ15	0.70 0.70	OC30	0.75
2N3134	0.80	A8Z16	0,70	OC35	0.50

0.80 0.25 0.70 0.70 0.75 0.25 0.125 0.125 0.30 0.175 0.15 0.15 0.15

0.20

0.25 0.40 0.25

0.20 0.20 0.30 0.40

0C28
0C29
0C30
0C35
0C36
0C42
0C44
0C45
0C71
0C72
0C78
0C78
0C78
0C81
0C81
0C81
0C81
0C14
0C141
0C171
0C20
0C171
0C200
0C200

OC201 OC202 OC203 OC204

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3-in tube fitted with telescopic viewing hood, giving bright display in full daylight. Sensitivity $10 \mathrm{mV/mm}$ (narrow band) to $30 \mathrm{nV/mm}$ (wide band). Bandwidth $10 \mathrm{c/s}-10 \mathrm{me/s}$. Triggered sweep pre-set at 1-2-5-10-30-100-300-1000-3000 μsec per atroke. Fre-running time base $20 \mathrm{c/s}$ to $2000 \mathrm{c/s}$ with built-in crystal calibrator providing timing marks at .05-.2-1-5-100.100 $\mu\text{-}$ crystal calibrator providing timing marks at .05-.2-1-5-1000 $\mu\text{-}$ crystal calibrator provid

20-100 \(\mu\) sec.

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0.60 0.60 0.75 0.50 0.20 0.20 0.15 0.10 0.12

0.30 0.20 0.20 0.25 0.20 0.25 0.15 0.20 0.30

0.80 0.85 0.60 0.25 0.25 0.80 0.60

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TYPE U4313. High sensitivity 20 Kohms/volt DC and 2 Kohms/volt AC. Accuracy 1.5% DC and 2.5% AC. 8 DC current ranges 60µA to 1.5A. 6 AC current ranges 0.60mA to 1.5A. 9 AC and DC Voltage ranges 1.5 to 500V. Resistance ranges 0.5-500 Kohms.

TYPE U4341 MULTIMETER AND TRANSISTOR TESTER



With taut band suspension move-

ment.

8-DC voltage ranges 0.3-900 V

6-AC voltage ranges 1.5-750 V

5-DC current ranges 0.06-600 mA

4-AC current ranges 0.3-300 mA

4-resistance ranges 0.5-500 Kohms

Transistor DC current gain 10-350

8ensitivity: 16700 Q/V DC:

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With taut band suspension move-

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5 AC current ranges 5mA to 2.5A. 7 AC/DC Voltage ranges 2.5 to

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Capacity range 0.5µr. 29-00

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2N3393 2N3394 2N3395 2N3402 2N3403 2N3404 2N3414

2N3414 2N3415 2N3416 2N3417 2N3702 2N3703

2N3704

2N3707

2N3709

2N3710

2N3710 2N3819 2N3906 28702 28746 AC113

AC125 AC126

0.75 0.30 0.80 0.20 0.15 0.15 0.20 0.15 0.15

0.20

0.25 0.25 0.12 0.12 0.17 0.15 0.12 0.12

0.35 0.20 0.50 0.25 0.15

0.80

A8Z16 A8Z17 A8Z18 A8Z20 A8Z21 BC107 BC108

BC108 BC109 BC113 BC118 BC134 BC147

BC148 BC149

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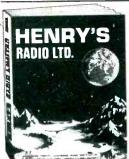
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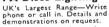
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